ideas for engineers

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63

301

February 1985

# RF Technology Expo '85 Issue

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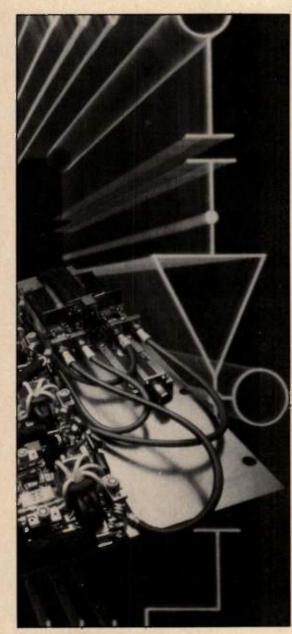
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#### February 1985



#### Special Report p. 26

#### Cover

#### February Cover —

The PB-2022 one-kilowatt amplifier provides 50 dB of gain in one modular "supercomponent." The PB-2022, and many other products, will be exhibited at RF TECHNOLOGY EXPO 85. Cover photograph courtesy of Acrian.

#### **Features**

36

#### 26 Special Report: Modules Lead the Way at RF TECHNOLOGY EXPO

Modular "supercomponents" can simplify life for the design engineer. Also a preview of the exhibits at RF TECHNOLOGY EXPO 85. Kiyoshi Akima.

#### The SAW Resonator: How It Works

The first part of this two-part series provides an introduction to the subject of surface acoustic wave resonators. A TI-59 program is given to model the responses of these devices. Jeff Schoenwald.

#### The Phase/Frequency Detector

46 This article discusses the design of phase-locked loops employing phase/frequency detectors. James Crawford.

#### Departments

- 6 Editorial
- 8 Publisher's Notes
- 11 Letters
- 16 News
- 61 RFI/EMI Corner
- 62 The Digital Connection
- 64 RF Designer's Notebook
- 67 Info Card
- 71 Product of the Month
- 72 New Products
- 88 New Literature
- 94 Advertisers Index

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### The Boom in RF Communications

rf editorial



#### Jim MacDonald Editor

A news item in last month's issue told about the sales a new California company anticipates in communication systems operating below 2 GHz. Many companies are expecting a growing market for RF communications equipment, especially overseas.

Much of the world depends on the RF spectrum for local communications. The Europeans have been using cellular radio for years, getting their components from American firms. *RF Design* has learned that one African nation has contracted with an American firm for a nationwide Rf telephone system.

We at *RF Design* expect to see an RF communications boom in the next few years to rival the computer phenomenon of the last 20 years. People in ordinary occupations are finding they need to communicate faster. Stock brokers in New York, for example, can no longer rely on the telephone or the "tape" for the kind of minute-by-minute stock and bond quotations they need in a volatile market. They are turning to radio as highly selective narrowband antennas make radio communication spossible in the canyon-like streets.

Even the military, the largest user of satellite communications, is a large-

volume buyer of RF equipment. Tactical battlefield communications and control systems still use VHF to a large extent, partly because short-range, line-of-sight communications are more secure.

Information exchange in the civilian sector is becoming increasingly timesensitive. The speed at which we can access information and perform calculations with digital equipment is mind-boggling, but how valuable is the information unless it can be communicated to the right people when they need it? America, with the best telephone system in the world, will probably continue to develop and improve satellite communications, but other nations may not have this capability for years.

Futurists talk about the decentralized workplace. They predict people will work at home on computers tied to a central office mainframe. If this does come about, think how important short-range radio and telephone communication may become. Electronic mail is no substitute for personto-person conversation, either for conveying complex ideas or for that contact with other people we all need.

We are flooded with information about the world, but we have little communication with each other. Most of us probably know more about the health of the latest Russian leader than we do about the health of our next door neighbor. Sociologists may know why we isolate ourselves — maybe it is necessary. The recent growth in amateur radio, however, and the tenacity with which CB radio has held on in spite of its problems indicate a deep desire to communicate with our fellow human beings, even when we have nothing important to say.

As this issue goes to press the *RF Design* staff is preparing for the RF Technology Expo, January 23-25, bringing together engineers, physicists and manufacturers in the RF field. We hope thi exposition and symposium will provide the foundation for a continuing exchange of information. *RF Design* will provide a forum for this ongoing exchange.

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### rf publisher's note

# **Giving Birth to a Community**



#### Keith Aldrich Publisher

As I write, the first RF TECHNOLOGY EXPO is still three weeks off...too early for a final count of attendees or exhibits or for a final assessment of its success. One thing is already clear, however, from the comments and attitudes of the people who have signed up in advance to be there. When closing time comes at noon on Friday, January 25, RF TECH-NOLOGY EXPO 85 will have given birth to an RF engineering community.

A community is defined as a group of people who have been drawn together by some common interest. It is the "together" that has been missing. Heaven knows, from the enthusiasm that has greeted every step of the development of RF TECHNOLOGY EXPO, that the common interest was already there. But RF engineers have never been able to congregate before and they have missed it. They have *needed* it. They have had to work in near isolation, even as their technology boomed.

We expect well over a thousand RF engineers at the Disneyland Hotel for the first RF TECHNOLOGY EXPO. That's no multitude, but it is a very good start of an influential and coherent community. We expect those engineers to go back to work renewed in their skills and rekindled in their zeal because of the association with their peers and counterparts from around the world. We expect their work to profit and a fresh infusion of design ideas and solutions. We expect new relationships to develop between people and companies that ought to know each other and will profit because they do.

A lot to ask? It's only the sort of thing that occurs in every strong community. And this one is just bursting to be born.

A word about my simile: "giving birth" to a community. Birth is generally accompanied by a lot of hard, even painful, labor. This one has been no exception, and I'd like to give credit to the main coordinators.

Kathy Kriner, manager of Cardiff Publishing's Trade Show Division, has brought a high degree of professionalism to the task of convention manager. I didn't realize (it's hard to imagine) how many details there are to be tended to in that chore and she hasn't missed any.

Andrzej Przedpelski, VP of Development for ARF Products, Inc., has given the event its essential character through his selection of the more than sixty papers that make up the technical conference. He has been tireless in what amounted to a second full-time job during a peakload period in his main one. *RF Design* will always be grateful.

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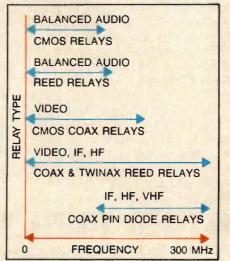
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30 to 900 MHz	BASE STATIONS							
Туре	Power (W)	wer (W) Freq. (MHz) Gain (db) min. VCC (V) Package						
BLW96	200	30	13 5	50	SOT-121			
BLV25	175	108	10 5	28	SOT-119			
BLV80 28	80	175	65	28	SOT-119			
BLV33F	85	225	10.5	28	SOT-119			
BLV36	120	225	100	28	SOT-161			
BLU53	100	400	65	28	SOT-161			
BLV97	30	860	65	24	SOT-171			
BLV57	38	860	65	25	SOT-161			

66 to 870 MHz	AMPLIFIER MODULES FOR LAND MOBILE					
Туре	Freq (MHz)	P In (MW)	P Out (W)	VCC	Package	
BGY32	68-88	100	20	12.5	SOT-132	
BGY33	80-108	100	20	12 5	SOT-132	
BGY35	132-156	150	20	125	SOT-132	
BGY36	148-174	150	20	125	SOT-132	
BGY43	148-174	150	13	125	SOT-132B	
BGY40A	400-440	100	75	12.5	SOT-132C	
BGY41A	400-440	150	13	125	SOT-132C	
BGY40B	440-470	100	75	12.5	SOT-132C	
BGY41B	440-470	150	13	125	SOT-132C	
BGY40A	470-512	100	7.5	12.5	SOT-132C	
BGY41C	470-512	150	13	12.5	SOT-132C	
BGY45A	68-88	150	30	125	SOT-301-A-03	
BGY45B	144-175	150	30	125	SOT-301-A-03	
BGY46A	400-440	30	15	96	SOT-26NC	
BGY47A	400-440	45	2.2	96	SOT-26NC	
BGY47B	430-470	45	22	96	SOT-26NC	
BGY47C	460-512	45	55	96	SOT-26NC	
BGY22	380-512	50	29	125	SOT-75A	
BGY23	380-480	2 5 WATTS	7	125	SOT-75A	





the mobile communications industry. These proposals call for channel spacings of as low as 7.5 KHz using various forms of single sideband modulation. (Channel spacing of 12.5 KHz are already common in Europe in the VHF bands.) With most signal generators only just having acceptable performance at 20 KHz carrier offset (the current channel spacing), it is debatable that these generators will be able to perform off-channel tests or be useful for development work on the proposed new systems. Synthesized signal generators such as the Racal-Dana Model 9087 or Hewlett Packard 8662 feature similar phase noise at 3 or 5 KHz carrier offset as at 20 or even 100 KHz from carrier.

Very close to carrier, offsets of less than 1 KHz, the phase noise becomes important when working with low deviation phase modulation and/or when multiplying to microwave frequencies. Again, it is only the multi-loop sophisticated signal generators that can meet these demanding requirements.

Most modern FM communications systems feature either digital data transmission modes or selective calling with digital coding. To evaluate these systems under all operating conditions requires the ability to frequency modulate the carrier with very low rates, or in some cases even DC. The lower cost synthesized signal generators often only use a single phase locked loop which requires closed loop DC operation to maintain frequency stability. These systems usually can only FM at rates above 20 or 50 Hz. The more sophisticated multi-loop systems available from some manufacturers allow true DC coupled frequency modulation.

Switching speed/setting time is perhaps the third important

area that may demand the consideration of a higher cost signal generator. For general purpose applications, a faster switching speed means that more tests may be made in a shorter period of time which can be important when, for example, fully characterizing a filter or checking a receiver for spurious responses. There are other applications particularly in the electronic warfare area that necessitate using a fast switching signal generator. Radio jammers often use a commercial synthesizer as the local oscillator. The response time of the jammer to track frequency agile systems often depends solely on the switching speed of the synthesizer. Switching speed is also important when checking the new generation of anti-jamming (ECCM) communication systems such as SINCGARS. To test an ECCM receiver dynamically, and with both traceability and repeatability requires that the signal generator can respond faster to a message to change frequency than the receiver. Currently, frequency switching speeds of less than 1 millisecond are required, which again precludes the use of the new economy signal sources.

I look forward to reading a further article on signal generators when the technology breakthroughs that influence performance rather than price/performance will be reviewed and discussed.

Sincerely, Malcolm Levy RF Products Manager Racal-Dana instruments, Inc. Irvine, CA 92714

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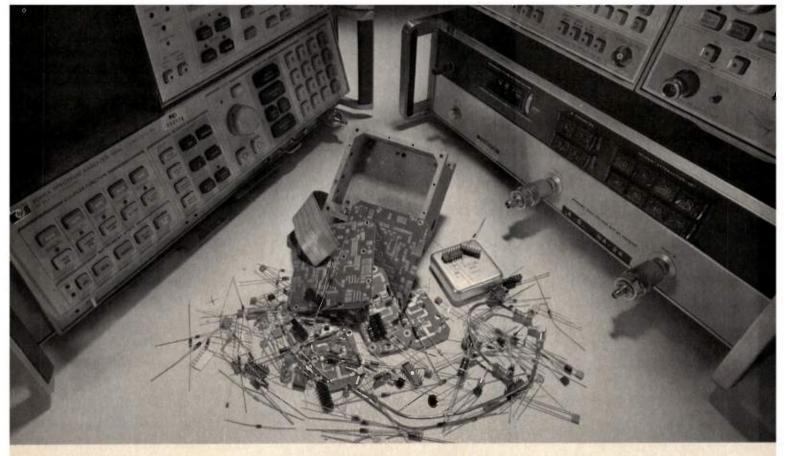
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### **rf** news

# RF Design Launches First RF Technology Expo

Engineers from the United States and 10 foreign countries were among the more than 1,000 registrants expected at the first RF TECHNOLOGY EXPO '85, Jan. 23-25, at the Disneyland Hotel, Anaheim, California. Sponsored by *RF Design* magazine, this was the first expo held specifically for RF design engineers. As this issue went to press, nearly the

As this issue went to press, nearly the entire RF design engineering staff from several companies were planning to attend RF TECHNOLOGY EXPO '85 (Motorola Semiconductor, Phoenix, Ariz., was planning to bring 30 engineers). An *RF Design* reader survey early last year showed many engineers lacked basic RF design knowledge or had forgotten what they once knew. Electrical engineers were being pressed into service in this rapidly growing field with little or no training. Reponse to the expo has verified the reader survey.

Manufacturers of RF devices and tools rented more than 75 booths to display their new and established products, and discuss them with design engineers. Survey respondents had expressed a need to learn about available design components and tools.

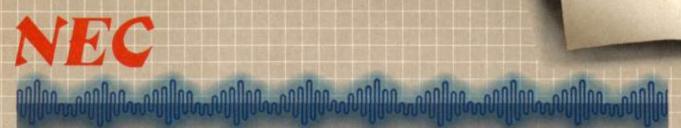
Most engineers registered for the expo design communications equipment, many for government, military, aerospace and EMC/EMI applications. Many others design instrumentation and test equipment for military and commercial use.

#### More than 60 Papers Presented

In addition to the display booths, these interests were reflected in more than 60 papers scheduled in 16 sessions during the three-day event.

To accommodate as many specific interests as possible, three sessions were scheduled to run simultaneously each morning and afternoon. Each session featured several papers related to a general topic, i.e., components, circuits or techniques. Attendees could learn about several aspects of each topic without changing rooms. *RF Design* will publish a Proceedings of papers presented in all sessions that may be purchased by attendees forced to choose between equally

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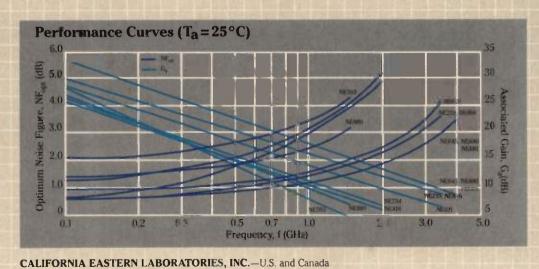


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INFO/CARD 17



attractive sessions and readers who could not attend RF TECHNOLOGY EXPO '85.

An all-day course, Fundamentals of RF Design, was scheduled for each of the first two days. Course instructors included Les Besser, president, Microwave Education Programs, Los Altos, Calif.; Professor K.C. Gupta, University of Colorado, Boulder; J. M. Johnson, president, Microwave Modules and Devices, Mountain View, Calif .; John Morton, engineering manager, Microsonics, Inc., Weymouth, Mass.; and Carl A. Erikson, Jr., director of processing operations, Anderson Laboratories, Bloomfield, Conn. RF Design is planning to include a transcript of this course in the Proceedings.

Encouraged by the response to RF TECHNOLOGY EXPO '85, RF Design is already planning next year's show. It, too, will be held at the Disneyland Hotel in January.



#### **RF Technology Expo '85 Exhibitors**

A.H. Systems, Chatsworth, Calif., EMI test equipment and antennas

Acrion, Inc., San Jose, Calif., Power transistors, products and assemblies

American Microwave Technology, Fullerton, Calif., OEM RF power amplifiers and sources

American Technical Ceramics, Huntington, New York, Porcelain and ceramic **RF** capacitors

Amperex Electronic Corp., Slatersville, Rhode Island, Wideband RF semiconductors and linear hybrids

Amplifier Research, Souderton, Penn., Amplifiers and accessories for RF and RFI susceptibility testing

Andersen Laboratories, Bloomfield, Conn., SAW oscillators and filters, frequency synthesizers and filter banks

ANZAC (Division of Adams-Russell), Dividers and log amplifiers

**Applied Engineering Products, New** Haven, Conn., Subminiature coaxial connectors and coaxial cable assemblies

Austron, Inc., Austin, Texas, Ovenized crystal oscillators and time frequency management equipment

Avantek, Inc., Santa Clara, Calif., Monolithic amplifiers, oscillators and mixer modules

Bird Electronic Corporation, Cleveland, Ohio, RF power measurement instruments, wattmeters and field-strength meter

Chomerics, Inc., Gardena, Calif., EMI shielding materials and components

Cirtech Corp., Stanley, Kansas, Frequency control products, crystal filters, etc. from DC to 200 MHz

Cohan-Epner Co., Booklyn, New York, Plated coatings for laser cavities and metal optics

Coil Craft, Cary, Illinois, RF fixed and tunable inductors and transformers

Communications Consulting Corp., Upper Saddle River, New Jersey, RF and microwave design and analysis software

Compact Software, Palo Alto, Calif., Software for design, analysis and optimization of MICs

Crystal Technology, Inc., Palo Alto, Calif., SAW bandpass filters, resonators, delay lines, convolvers and VCOs

Dow-Key, Santa Barbara, Calif., Electromechanical coaxial RF/microwave switches

EEsof, Inc., Westlake Village, Calif., RF/microwave software for design and analysis: Touchstone and MICAD

**Electronic Surveillance Components,** Inc., YIG Tuned transistor oscillators and solid state HP-8551 BWO replacement

**EMC Technology, Inc.,** Cherry Hill, New Jersey, SMA connectors, terminations, loads, attenuators and MICs

ERBTEC Engineering, Inc., Boulder, Colo., Computer-enhanced HF CODAR radar system, amplifiers and time systems

Frequency Sources, Inc., Chelsmasford, Mass., Semiconductor products, varactors, diodes and capacitors

Greenray Industries, Crystal oscillators, VCOs, temperature compensated oscillators

Hewlett-Packard, Spokane, Wash., Signal generators and phase noise measurement equipment.

Holaday Industries, Inc., Eden Prairie, Minn., Isotropic broadband field strength meters with self-zeroing fields.

Intech, Inc., Santa Clara, Calif., Communications and satellite navigation equipment

JFW Industries, Inc., Indianapolis, Indiana, Fixed and programmable attenuators, divider and detectors

Johnson Manufacturing Corp., Boonton, New Jersey, Variable capacitors and microwave tuning elements

K & L Microwave, Inc., LC and cavity filters, phase meters and distorition analyzers

K & L Quartztek, Phoenix, Ariz., Quartz crystal filters and oscillators

Kay Elemetrics Corp., Pine Brook, New Jersey, RF attenuators

Keene Corp./Ray Proof Div., Norwalk, Conn., RF shielded rooms, anechoic chambers, absorbers and filters

L & M Engineering South, Inglewood, Calif., RF and microwave components and subsystems

Logimetrics, Inc., TWT amplifiers and RF signal generators

Marconi Instruments, Allendale, New Jersey, RF power meters, microwave scalar analyzer and signal generators

Matrix Systems Corp., Calabasas, Calif., Coaxial and audio switching systems and modules

Microwave Research Corp., RT/duroid microwave materials

Microwave Modules and Devices, Mountain View, Calif., High-power, solid state microwave modules, including amplifiers

Microwave Semiconductor Corp., Somerset, New Jersey, A full line of RF/microwave semiconductor products and components

Motorola, Inc., Franklin Park, Illinois, Quartz crystals, filters, oscillators and resonators

Motorola Semiconductor, Tempe, Ariz., RF transistors — module and discrete, linear products, logic functions

P & H Laboratories, Inc., Ferrite

isolators and circulators, isoadapters and wave guide assemblies

Pacific Research and Development, Phase locked oscillators and frequency multipliers

Polarad Electronics, Inc., Lake Success, New York, RF test and measurement equipment.

Proto Stamping Corp., Redwood City, Calif., RF/EMI shielding and metal fabrication Racal-Dana Instruments, Inc., Irvine, Calif., RF test and measurement equipment

Rogers Corp., Rogers, Conn., RT/duroid microwave materials

SAWTEK, Orlando, Florida, SAW devices

Soladyne, Inc., San Diego, Calif., Stripline and microstrip circuits

Swift and Associates, Van Nuys, Calif., Swift wrench and other products

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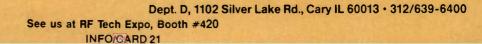
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#### The Program in Brief

Wednesday, Jan. 23, 8:30 a.m.-noon

Fundamentals of RF Design — Part 1

#### **Computer-Aided Design**

"Use of personal computers as a tool in prediction and control of EMI" "CAD methodology for microwave

oscillators" "Small computers in RF design"

"CIAO and design: circuit optimization and synthesis programs for personal computers"

#### **Components** — Transistors

"Hybrid varactor-tuned oscillator modules: their practical applications in RF communications"

"Surface mounted components" "Integrated circuits for IF amplifiers and demodulators"

"Motorola advanced amplifier concepts"

#### **Transmitters**

"A 4kW broadband (40 kHz to 40 MHz) pulsed power amplifier"

"RF operation of 450 volts vertical power MOS transistors in an ultra-lightweight HF transmitter"

"Low noise UHF transmitter design"

Wednesday, Jan. 23, 1:30 p.m.-5:00 p.m.

#### Fundamentals of RF Design — Part 2

#### Circuits

"Design compromise in single loop frequency synthesizers"

"A temperature stabilized RF power detector"

"An ultra-fast UHF voltage controlled attenuator with 35 dB linear dynamic range" "Design of GaAs power FETs module"

#### Techniques

"Mounting considerations for RF low power plastic packages"

"Using Avantek MODAMP MMICs in broad and narrow-band filter design" "Computer-aided design of phaseback loop circuits"

#### Antennas

"Helical antenna design" "Snyder antenna" Thursday, Jan. 24, 8:30 a.m.-noon

#### Fundamentals of RF Design — Part 1

#### Components

"The hybrid amplifier module for cellular telephone design"

"High frequency blind mating connectors for efficient microwave packaging" "High voltage UHF power static induction

transistors"

#### Circuits

"Aspects of discriminator design" "A precision glitch-free RF step attenuator"

"A 900 MHz super amplifier" "Precise bandwidth combline filter

realization"

#### **Techniques**

"Amateur satellite communication uplink" "Testing digital data systems"

"Precision phase noise measurements of oscillators and other devices from 1 MHz to 20 GHz"

"Adaptive HF radio systems"

Thursday, Jan. 24, 1:30 p.m.-5:00 p.m.

#### Fundamentals of RF Design — Part 2

#### **Computer-Aided Design**

"Matching network design using HP-41, HP-71 and HP-75 computers" "Computer and calculator aided design

tools for the RF engineer"

"RF computer aided engineering — mainframe performance on microcomputers"

#### **Transmitters**

"Large signal power amplifiers" "A 2 GHz amplifier provides telemetry

capability"

"An integrated Ka-band power amplifier" "Class-D power amplifier load impedance for maximum efficiency"

#### Techniques

"Application of digital signal processing" "ACSB — an overview of amplitude compandored sideband technology" "Electrical characterization of quartz

crystal resonators through high performance vector network analysis" "Application of coherent digital memories of communications jamming" Synergy Microwave, Patterson, New Jersey, Mixers, power dividers modulators, quadrature hybrids and transformers

Tektronix, Inc., Woodland Hills, Calif., Spectrum analyzer, tracking generator and accessories

Tele-Tech Corp., Bozeman, Montana, Signal processing components and RF/ microwave subassemblies

Telonic Berkeley, Inc., Laguna Beach, Calif., RF and microwave filters, RF attenuators and DSWR systems

Texscan Corp., Phoenix, Ariz., Spectrum analyzers, sweep generators, synthesized signal generators

Trilectric, Inc., Las Vegas, Nevada,

TRW RF Devices Division, Lawndale, Calif., Low noise small signal transistors, other transistors and RF amplifiers

Vari-L Co., Denver, Colo., Wideband signal processing components, mixers, transformers and hybrids

Varlan-Eimac, San Carlos, Calif.,

Varitech Electronics, Inc., Lynbrook, New York, Fixed RF capacitors, dialectric and variable capacitors

Vectron Laboratories, Inc., Norwalk, Conn., Crystal oscillators, including TCXOs, VCXOs, VCOs, OCXOs and hybrids

Voltronics Corp., East Hanover, New Jersey, Precision trimmer capacitors

Wavetek, San Diego, Calif., Microwave and RF miniature components, RF signal sources

#### "Micro-Coax" Gives Its Name to New Company

A new company, Micro-Coax Components, Inc., has been formed by UTI Corp., Collegeville, Penn., out of its former Micro Delay Division in Collegeville, effective January.

Growing along with the booming RF market, the Micro Delay Division of UTI has had to expand its facilities four times in the past decade. The continuing growth has "dictated incorporation as a separate subsidiary," said Dr. Robert H. Schafer, president of the new Micro-Coax Components.

The Micro Delay Division was established in 1962 by Uniform Tubes (also a division of UTI) to produce miniature semirigid coaxial cable. The division developed several advanced techniques for producing a high quality, reliable semirigid coaxial cable and gave it the name Micro-Coax.

The Micro-Coax brand name's reputation was a major factor in selecting the name Micro-Coax Components, Inc., for the new operating company. Micro-Coax is the holder of patents for several other

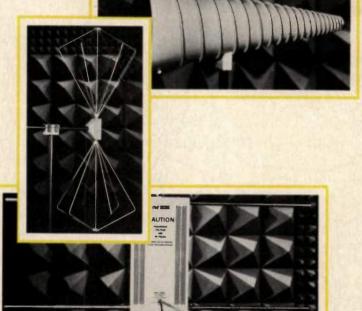
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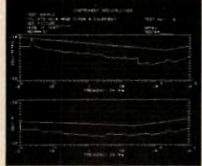
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#### Thews Continued

products, such as impedance transformer cable and assemblies, and low-pass, bandpass and high pass filters implanted within the cable assemblies. The "In-A-Cable" filters are being widely used for the suppression of undesired signals in transmitting sources and suppression of unwanted frequency bands in receiving systems. The company also offers a line of coaxial connectors, and coaxial and waveguide delay lines.

Micro-Coax Components, Inc., is planning to introduce several new products during 1985, according to Dr. Schafer.

#### EG&G/Cinox Pact Gives New Depth in Oscillators

Cinox Corp., a Cincinnati, Ohio firm familiar to *RF Design* readers as a maker of quartz crystals and crystal-based oscillators used for frequency control and stabilization, has been acquired by EG&G Inc. of Wellesley, Mass., for an undisclosed amount of cash.

Cinox has been since 1930 a privatelyheld company and now employs more than 200 people. Commercial and military applications of its components are primarily in the areas of communications, navigation, guidance, tracking and telemetry.

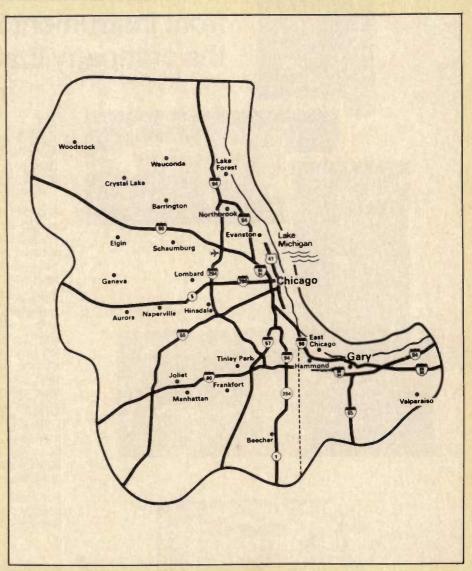
Cinox President W. Kirk Dance said the Cinox expertise in the design and manufacturer of high-performance oscillators would complement that of EG&G's Salem, Mass.-based Electronic Components Division in the development of atomic frequency standards for future military and commercial voice and data communication systems; space and global navigation systems; and military command, control and intelligence systems.

"The acquisition makes sense," Dance said. "Their technology matches our technology." He said the acquisition provides Cinox with the capital backing to increase an already impressive rate of growth. "We have grown so rapidly in the past four years that it attracted EG&G's attention," Dance said. "They have more capital than we do so we can grow faster."

In a private interview with *RF Design* (the phone is now answered "EG&G Cinox"), VP of Marketing Ron Stephens alleged that new product developments traceable to the new connection would be announced "shortly." For more information contact Ron at (513) 542-5555.

#### Cells Multiply In Chicago Carphone Market

Ameritech Mobile Communications, Inc., operator of the nation's first commercial cellular carphone service — launched



Cellular Service coverage for the Chicago/Gary area.

in Chicago in October 1983 — recently announced plans to double the size of the Chicago system by mid-1985. The initial service area covered 2,500 square miles in the greater Chicago metropolitan area, ranging from Lake Forest on the north to Geneva on the west to Beecher on the south. In the announced expansion plan the coverage area will increase to 5,000 square miles, and range from Waukegan on the north to Gary, Ind., on the south (see map illustration).

Ameritech Mobile, which introduced its service with just 17 "cells," has since increased the number of cell sites to 29 and with the new expansion will boost that number to 56, including seven new cells specifically designed to improve service in downtown Chicago. The expansion will double customer capacity as well as system size. The company currently has about 12,000 customers. Capital investment to date has totalled \$45 million. The expansion is expected to require another \$21 million, according to Ameritech.

Most of the capital invested goes toward the purchase of RF equipment. Each cell site in the 56-cell system requires a transmitting/receiving tower and remote switching gear. Car telephone radios, receivers and antennas represent additional investment on the part of individual subscribers.

Ameritech Mobile, a subsidiary of Bell's Ameritech, also announced the availability of new custom calling features to enhance the carphone service. Call waiting, call forwarding, three-way calling and busy/no answer transfer are being introduced now to current customers. A revised tariff schedule filed with the Illinois Commerce Commission last October 24 reflects generally falling prices for standard services (e.g., \$20 one-time charge to establish service).



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In low-power, the U.S.-made, GaAs MRF 966 provides 1 dB gain at a gigahertz, the MRF 557 PowerMacro 1.5 W output in popular plastic packaging and the MRF 571 7-8 GHz fT performance from the MRF 571.

#### Super carrier for marine (SSB) radio.

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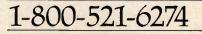
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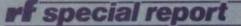
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# Modules Lead the Way at RF TECHNOLOGY EXPO

By Kiyoshi Akima RF Design

Many of the exhibitors at RF TECH-NOLOGY EXPO are displaying modular devices that can be used as "supercomponents." Very few engineers will take the time to design a component, for example a resistor, needed in a system. One simply specifies the desired performance, and purchases a resistor. In the same way, a designer can specify a device such as a power combiner or an amplifier, and purchase a module which can then be used as if it were a component.

Both active and passive devices are available as modules, and they are available as monolithics or hybrids. They may be constructed using either thin or thick film technology, or both. While the remainder of this article focuses on one of the most widely available types of modular devices, the power amplifier, most of the discussion is also applicable to other types such as attenuators and frequency mixers.

Each type of construction has its advantages and disadvantages. Monolithics are formed on a semiconductor substrate by deposition and photolithography, as a printed circuit. As a result, they are generally smaller physically and less expensive than the hybrid modules. They can require less than a square inch of board space, and cost under \$2 in production quantities.

Hybrids are constructed with discrete components on substrates. At the expense of size and cost, they provide better performance and higher power levels than can be achieved with monolithics. RF power amplifiers producing over 100 W are available, but they can cost as much as \$200 or more.

Being standard 50 ohm in/50 ohm out devices, modules can be used without the need for additional mixers and shunts. If a single module won't do the trick, more can be cascaded, within the limits of each unit's power rating. Most modules do not allow external adjustments and tuning.

Modules provide many advantages. Among them are repeatability, size, time, and cost. The variations in component values have been taken into account, and the modules come "pre-tweaked" to meet tight tolerances. This can reduce the amount of adjusting and tuning required for each circuit.

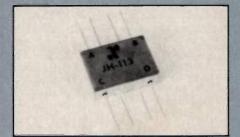
By using modules instead of discrete components, the resulting circuit can usually be made much smaller and simpler. Fewer components mean fewer interconnections. Fewer connections mean more leeway in the layout. Often the heat dissipation of a module is less than that of an equivalent discrete circuit. This can eliminate additional heat sinks or cooling systems.

The use of modules also considerably reduces the time required to produce a product. Once the requirements for an amplifier have been determined, the amplifier does not have to be designed and developed. The module manufacturer has already done this for you. The module, by being one component instead of a collection of elements reduces the time required to assemble the circuit. By not needing to tune each circuit the assembly time is reduced further.

Cost is another item that can be reduced by using modules. The time savings in the design and manufacturing phases translate directly into cost savings. Smaller circuits require less materials, and can be housed in smaller enclosures. Sometimes, the modules themselves also costs less than the discrete components required to build the equivalent circuit.

#### Supercomponents

The Anzac Division of Adams-Russell has developed a new quadrature hybrid to cover the 7 to 14 MHz band in a miniature flatpack. The JH-113 features typi-



Adams-Russell

cally 27 dB isolation and a low loss of 0.5 dB over the entire octave. Other features include 1.2:1 VSWR and 0.75 dB amplitude balance. The JH-113 is a hermetically sealed unit for military environments. Adams-Russell Co., Anzac Div., Burlington, MA 01803. Booth #321, 323, 325; Info/Card #112.

The Avantek MTO-8040 varactor-tuned oscillator covers 400 to 600 MHz with +10 dBm power output into a 50 ohm load. It



#### Avantek

is packaged in TO-8V transistor cans and is fully tested and guaranteed over the full -54 to 85 C military temperature range. Avantek, Inc., Santa Clara, CA 95051. Booth #516, 518; Info/Card #111.

The BT500 active bias thermal tracking device from TRW provides excellent thermal tracking and requires only limited external circuitry for operation. It is available in a variety of packages, including the standard hermetic TO-60. TRW Electronics Components Group, RF Devices Div., Lawndale, CA 90260. Booth #217, 219; Info/Card #110.

The Watkins-Johnson WJ-EA54 Economy Amplifier provides 27 dB typical gain from 10 to 250 MHz. The unit's noise figure is 3.8 dB and its output power the (Continued on page 29.)



Watkins-Johnson

A 1000 watt (1 kilowatt) solid state amplifier, the Acrian PB-2022, is designed for military and industrial communications, and research equipment. It is characteristic of a family of standard wideband high power amplifiers available from Acrian, Inc.

As the industry has grown in size and made advances in technology, competition has increased dramatically. Combined with a shortage of RF and microwave design engineers, it has become essential for systems manufacturers to shorten design time for customers. This has resulted in more and more component suppliers offering modules and sub-assemblies to systems houses. It saves customers substantial time and money in a rapidly moving market.

Custom sub-assemblies, like the PB-2022, are designed to meet these needs. Because they are familiar with advances in device technology, engineers at companies such as Acrian can select current state-of-the-art devices for wideband power amplifiers and therefore provide a technology edge for customers. Well-designed amplifiers can then be easily adapted to a standard product line, allowing systems companies faster access to markets.

#### The PB-2022

Designed to meet military specifications — like all Acrian sub-assemblies — the PB-2022 has wide instantaneous bandwidth (30-150 MHz), high power gain (+50 dB), ALC gain flatness, plus wide dynamic range (60 dB). It is unconditionally rugged and stable at 3:1 VSWR and has a wide temperature range (ambient to +50 C, with heat sink to +70 C). Offering superior performance, the PB-2022 uses Acrian's proprietary ISO-FET device technology.

As with all other Acrian subassemblies, the PB-2022 can be



repaired in the field. It can be easily removed and replaced to minimize down time should a failure occur. The input splitter and the output combiner have excellent port-to-port isolation. This is important in the event of a failure in the output stage. If this happens, the amplifier will continue to operate at a reduced power level.

Because the PB-2022 is so efficient, a simple heat exchenger can be used. The amplifier can be easily cooled with only conventional heatsinks and forced air. When mounted into a 19 inch by 7 inch high rack, the unit can be cooled by using a 300 CFM of air with a back pressure of 1 inch of water with the proper selection of heatsinks and mechanical layout.

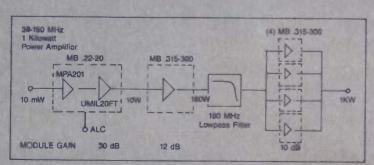
#### **Design Features**

The PB-2022 is made from two basic 50 ohm power block modules. The first is a 20 watt pre-amplifier, similar to the Acrian MB22-20. Two stages of amplification are employed: the first is a 50 ohm hybrid followed by a power FET. An ISO-FET was used at this point in the line-up due to its superior gain versus controlled voltage characteristic as well as its 4 dB noise figure and 15 dB of gain. More than 60 dB dynamic range is achieved in this single power block.

A common 300 watt module is used for the driver and in the combined output stage. Extensive use of ferriteloaded coaxial transformers assure multi-octave bandwidths. The input has a 4:1 ratio, while the output has a 9:1 ratio. The quiescent current of each module may be individually set with an on-board potentiometer, thereby allowing module swapping. Typical module efficiency is greater than 60 percent.

The transistor is composed of 2 matched pills mounted on a common flange and is operated in the push-pull mode. This is called the "Gemini configuration." Both even and odd mode terminating resistors assure stable operation.

A conventional 50 ohm, 4-way inphase splitter and combiner is used in the PB-2022. Unique combinations of 2-way, 6-way, and 2 x 4-way configurations have been demonstrated — all using the basic 300 watt power block module. Acrian, Inc., San Jose, CA 95126. Booth #301, 303; Please circle INFO/CARD #189.



Specifications Operating Frequency Power Output Power Output into 2:1 VSWR Power Input Power Gain ALC Control Input Impedance Operating Voltage Efficiency Overall Harmonics Stability

30 to 150 MHz Instantaneous 1 KW into 50 ohm Load 500 Watts Minimum +10 dBm 50 dB Minimum Output Level Control Provision 50 ohms 2.0:1 VSWR Maximum 28 Volts 40% Min. 2nd <-25 dBc/3rd <-15 dBc Unconditionally Stable in 3:1 VSWR

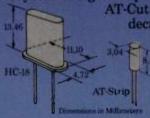
# AT-STRIP QUARTZ CRYSTALS SET NEW SIZE AND PERFORMANCE STANDARDS.



#### 90% SMALLER THAN HC-18 WITH GREATER SHOCK RESISTANCE AND A WIDE FREQUENCY RANGE

Motorola introduces the latest technology in crystal miniaturization. This success is part of our ongoing commitment to reliability and smaller size.

It is the 3.0 mm x 8.4 mm AT-Strip Quartz Crystal. You get the performance of conventional



AT-Cut Quartz in a significantly decreased package volume. Both new and existing applications benefit from this development – pocket pagers, disk drives, modems, commercial communi-

cations devices, and driving microprocessor-based systems.

#### WITHSTANDS 8,000 g SHOCK FORCE

The new Motorola AT-Strip Crystals offer a shock survival level two and three times higher than traditional crystals.

These and other environmental characteristics are documented in tests performed in accordance with MIL STD 202. TYPICAL CRYSTAL SHOCK PERFORMANCE

#### Our advanced

2000

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turing and quality assurance capabilities provide unsurpassed reliability in the field.

#### EXPANDING AVAILABILITY OF FREQUENCIES

Motorola AT-Strip Crystals are currently available in popular fundamental frequency ranges. And our engineers are adding frequencies

daily – with the shortest lead times in the industry – to meet your needs.

That means you can start incorporating a new AT-Strip Crystal in your



applications now. Give your new products an edge. And improve your current product's performance. Contact us for complete literature on the full line of Motorola Devices.

Motorola, Inc. Components Division, 2553 N. Edgington Street, Franklin Park, IL 60131. Phone (312) 451-1000, Ext. 4414 TWX 910-255-4619, Telex 4990101.



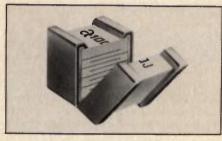
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1-dB compression point is +5 dBm. Typical output VSWR is 1.2:1 and DC current required at 5 v is 30 mA. Watkins-Johnson Co., Components Applications Eng., Palo Alto, CA 94304. Booth #500, Info/Card #109.

#### Components

A new termination metallization called UNI-TERM has been developed by American Technical Ceramics. The process uses silver or silver palladium thick film as the primary layer. A proprietary nickel layer is then applied to the thick file material. A final application of a fugitive layer of pure gold completes the process. Some of the advantages of UNI-TERM are: compatibility with all standard circuit



American Technical Ceramics

attachment techniques including wire bonding, soldering, and epoxy bonding; excellent solderability; and reduced termination material migration. UNI-TERM is available at no extra cost on any ATC chip, and will be standard on new products. American Technical Ceramics, Huntington Station, NY 11746. Booth #611; INFO/CARD #176.

Surface acoustic wave hybrid oscillators from Andersen Laboratories are available in a standard design package for a wide range of applications. Compact,



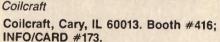
#### Andersen Laboratories

low phase noise devices requiring no alignment, they are available with operating frequencies from 100 MHz to greater than 1 GHz. Both fixed frequency and VCO devices are available. Andersen Laboratories, Inc., Bloomfield, CT 06002. Booth #211; INFO/CARD #175.

A family of 21.4 MHz IF filters packaged in less than ½ cubic inch (1.75"L x 0.6"W x 0.44"H) will be exhibited by Cirtech. The family is available in 3 dB bandwidths from 2 kHz to 10 MHz and features spurious responses of better than 60 dB, attenuation better than 70 dB, and ripple below 1 dB. The size uniformity is ideal for applications offering optional bandwidths. The 0.44" height allows half-inch board spacings. These filters conform to MIL-STD-202, Method 204, Condition B. **Cirtech Corp., Stanley, KS 66223. Booth #511; INFO/CARD #174.** 

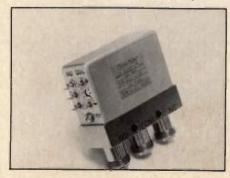
Coilcraft is exhibiting their line of surface mount inductors and transformers.





Crystal Technology's surface acoustic wave filters can be used for IF filtering, timing extraction, and general spectral shaping in a wide variety of applications, including TV and CATV, satellite receiver, telecommunications, radar, and telemetry systems. They provide greater selectivity, reproducibility, and stability than either LC or helical filters, especially at UHF or L-band frequencies. Out-of-band rejection of 70 dB can be realized with a single filter, and over 100 dB can be achieved by cascading two units. Crystal Technology, Inc., Palo Alto, CA 94303. Booth #412; INFO/CARD #172.

The Dow-Key 402 series SPDT "Smart Switches" use the latest MIC detector and solid-state CMOS driver technology internally to detect power and VSWR changes and switch automatically when preset conditions are met. They feature high power handling (350 W at 1 GHz) and low insertion loss (0.5 dB at 12 GHz). The



Dow-Key Microwave

"Smart Switches" are designed for redundant transmitter applications where automatic switching is required if the power decreases in one channel (amplifier failure) or the antenna pattern is disturbed (disconnected or damaged). Dow-Key Microwave Corp., Carpinterla, CA 93013. Booth #105; INFO/CARD #171.

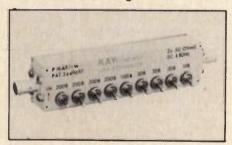
The TSO500 miniature thin film chip attenuators are available from EMC Technology in 1 to 20 dB values, for the frequency range from DC to 18 GHz. At-



EMC Technology

tenuation accuracy is  $\pm 0.25$  dB. Impedance is 50 ohms and power dissipation is 0.1 W. The attenuators operate from  $-55^{\circ}$  to 150°C. Measuring only 0.075" x 0.060" with 0.025" terminals, these automatically trimmed attenuators are suitable for circuits of all types where packaging density and improved high frequency performance are important considerations. EMC Technology, Inc., Cherry Hill, NJ 08034. Booth #309; INFO/CARD #170.

The Model 439 from Kay Elemetrics is a miniature bench attenuator designed to provide a frequency range of DC to 1500 MHz. Attenuation range is 0 to 101 dB in



#### Kay Elemetrics

1 dB steps. VSWR is 1.2 below 500 MHz and 1.4 above 500 MHz. Insertion loss is 0.2 dB below 500 MHz, 0.5 dB between 500 and 1000 MHz, and 0.8 dB above 1000 MHz. The 439 is available with BNC connectors, with SMA or TNC available as options. Kay Elemetrics Corp., Pine Brook, NJ 07058. Booth #423; please circle INFO/CARD #169.

The "A" and "E" line of internally sealed air trimmers offer ten full turns of adjustment, long life, Qs over 5000, internal stops, extended shaft options, and tuning ranges of either 0.8 to 10 pF or 0.8



Voltronics

to 14 pF. Their non-rotating piston glass trimmers offer much higher capacitance values, exceeding 250 pF. The non-rotating design provides excellent linearity, high current capacity, low RF loss, high Q, and long life. The "M" series of sapphire trimmers are usable to 3 GHz while the "CP" series of surface-mounted trimmers are usable up to 5 GHz. Voltronics Corp., East Hanover, NJ 07936. Booth #613; INFO/CARD #168.



High quality attenuators don't necessarily have to include high quality prices. Kay attenuators are designed and manufactured with high quality components but we do not subject our buyers to high prices. The Model 439 was designed to offer all of the superior specifications of the Model 432D but at a reduced cost, providing a \$36.00 savings to the customer. Kay is able to produce low priced attenuators with superior specification by using teflon, gold and silver component in our custom developed and in-house manufactured switches. Listed below are some of the more common attenuators. Check the prices for yourself.

ATTENUATOR TYPE	MODEL NO.	IMPED- ANCE	FREQ. RANGE	ATTEN	STEPS	INSERTION LOSS AT 1GHz	PRICE*
Standard In-Line	432D 442D	50Ω 75Ω	DC- 1GHz DC- 1GHz	0-101dB 0-101dB	1dB 1dB	.7dB .4dB	\$265 \$265
Miniature In Line	439A 449A	50Ω 75Ω	DC-1.5GHz DC- 1GHz	0-101dB 0-101dB	1dB 1dB	.5dB .4dB	\$229 \$229
Rotary Rotary Bench:	500A 510A 5050	50Ω 75Ω 50Ω	DC · 2GHz DC · 1.5GHz DC · 1GHz		1dB 1dB .1dB	.2dB .2dB .8dB	\$175 \$175 \$589
Programmable	4440 4457	50Ω 75Ω	DC-1.5GHz DC- 1GHz		10dB 1dB	2dB 3dB	\$299 \$375

\*Single Quantities. Discounts for Quantity Orders.

If you don't see an attenuator that fits your requirements call us for information on our complete line of standard products. Kay can also design an attenuator to fit your specific needs. We have designed over 250 different special attenuators in the past.

We also offer a very substantial discount schedule for quantity purchases. For a complete catalog or to place an order call Chris Meagher at (201) 227-2000, ext. 105.



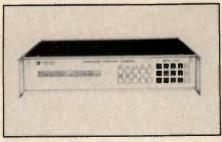
Kay Elemetrics Corp 12 Maple Avenue Pine Brook, NJ 07058-9797 USA

Tel: (201) 227-2000 TWX: 710-734-4347

INFO/CARD 27

#### Systems

The Austron Model 2110 is a microprocessor-controlled disciplined frequency standard that automatically locks the frequency of its precision ovenized crystal oscillator to that of an externally applied reference having superior long-term stability. Using a third-order servo tech-



Austron

nique, the instrument is able to correct the frequency offset and aging of the internal oscillator. If the external reference is removed or fails, the 2110 will continue to apply corrections to the oscillator to hold it on frequency. Typically the unit can limit the frequency offset to a few parts in 10<sup>12</sup> for several days following the loss of the reference. Austron Inc., Austin, TX 78761. Booth #602; INFO/CARD #167.

Designed to be part of a larger multikilowatt RF power amplifier system, the Erbtec RF Amplifier Controller incorporates the intelligence and all the necessary electronics to monitor, control, and properly drive external high power vacuum tube amplifiers. Its modular architecture includes a RS-232 interface enabling all functions to be completely controlled and monitored by a remote computer. The instrument has very few manual adjustments, and no special tools are required for maintenance. Erbtec, Inc., Boulder, CO 80301. Booth #316; INFO/CARD #166.

Using MOSFET technology, the Intech COM 1000 provides 1000 watts over a frequency range of 1.6 to 30 MHz. The harmonic distortion is at least 73 dB below the fundamental at rated output power. The COM 1000 incorporates a "soft fail" mode in which operation is impaired but not inhibited upon failure of one or more FETs. In this mode the output is less than 3 dB down. Intech, Inc., Santa Clara, CA 95050. Booth #502; INFO/CARD #165.

#### **Test Equipment**

A solid-state, cavity-tuned, RF power signal source, the RPS 3000 from American Microwave Technology produces 10 watts of CW output from 3.0 to 3.5 GHz. The output is isolated to safely provide maximum power operation into any VSWR. Actual power output can be monitored on the front panel. American Microwave Technology, Inc., Fullerton, CA 92631. Booth #520; please circle INFO/CARD #164.

A new series of five plug-in elements from Bird Electronic converts the company's THRULINE wattmeter into an extraordinarily sensitive instrument. The elements offer bi-directional RF power measurement from 20 mW to 100 W at



**Bird Electronics** 

 $\pm 5\%$  accuracy of reading, in frequency bands from 25 to 1000 MHz. Full rated accuracy is maintained over the 5000 to 1 power range with temperature extremes between 0° and 50°C. The entire complement of elements enable the THRULINE to cover frequencies from 200 kHz to 1000 MHz and power from 20 mW to 10 kW. **Bird Electronic Corp., Solon, OH** 44139. **Booth #307; INFO/CARD #163**.

Four surface acoustic wave resonator oscillators form the heart of the Hewlett-Packard 8642A/B signal generators. These extremely high-Q devices operating near 800 MHz provide low SSB phase noise at offsets beyond 10 kHz. The 8642A covers the range of 100 kHz to 1057.5 MHz while the 8642B doubles the upper limit to 2115 MHz. Both models offer underrange to 10 kHz and a resolution of 1 Hz, 0.1 Hz with a special function. The



Hewlett Packard RF Design output ranges from +20 to -140 dBm with a resolution of 0.1 dB. AM, FM,  $\phi$ M, and PM allow the 8642A/B to accurately simulate signals used for most types of RF communication systems. On demand, the generators can switch in a separate heterodyne output section to provide improved modulation performance for carrier frequencies below 132.2 MHz. Hewlett-Packard, Palo Alto, CA 94303. Booth #522; INFO/CARD #161.

Weighing in at 7.5 kg, the Marconi 2022 signal generator has a frequency range of 10 kHz to 1000 MHz, with a resolution of 10 Hz below 100 MHz and 100 Hz above 100 MHz. Output amplitude is +6 to -127 dBm with a resolution of 0.1 dB. Amplitude, frequency, and phase can be modulated externally or internally. Two elapsed time indicators — usage since last resetting and usage since manufacture — aid in analyzing reliability and cost-



# SAW Oscillators

Sawtek's Surface Acoustic Wave oscillators for military and commercial applications simplify design and improve noise performance. High-Q SAW resonators offer quartz stability at fundamental frequencies from 100 MHz to 1000 MHz. Hybrid oscillators in hermetic packages are available for reduced size and increased reliability. FM or pulse code modulation capability is optional.

Sawtek maintains a large selection of frequencies from an inventory of pre-tooled resonator crystals and new designs can be tooled rapidly. Our engineers also offer assistance in oscillator design for low-cost consumer applications and are prepared to help evaluate the suitability of SAW oscillators for your requirements.

In addition to oscillators and resonator products, Sawtek produces other high performance SAW components including bandpass filters, delay lines, and pulse compressors for cable television, satellite communications, moderns, radar, EW, and many other signal processing applications. And, if what you need is not among our hundreds of standard products, we can provide technical assistance and rapid response to new design and production requirements. Quality and performance have made Sawtek the industry leader in SAW technology; you can rely on us for the total engineering support you need.

When your system demands the advantages of SAW Technology ... Demand SAWTEK.



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INFO/CARD 28

#### of-ownership. Marconi Instruments, Allendale, NJ 07401. Booth #507; please circle INFO/CARD #160.

For testing of communications systems, one of the most important parameters for a singal generator is its phase noise performance. The new FCC proposals to reduce channel spacings to as low as 7.5 kHz place stringent requirements on instruments used to test the new systems. The Racal-Dana 9087 generates signals with noise as low as -142 dBc/Hz at only 3 kHz offset from the carrier. With a frequency range of 10 kHz to 1300 MHz with a resolution of 1 Hz, the 9087 covers all HF, VHF, and UHF telecommunications bands as well as the LF and MF bands often used for radio navigation. The signal may be swept. The output is variable from +19 to -140 dBm (2 V to 0.0224 µV rms into 50 ohm) with a resolution of 0.1 dB. A GPIB interface provides for full programmability. In addition, its fast switching speed of 500 µs makes the 9087 suitable for FSK (frequency shift keying) simulation and the testing of electronic warfare systems. The 9087 offers total modulation



#### Racal-Dana

versatility, including AM, FM,  $\phi$ M, and PM from a combination of internal and external sources. Simultaneous AM+FM, AM+ $\phi$ M, FM+PM, and  $\phi$ M+PM allows accurate simulation of ECM/ECCM and "chirp" signals. Racal-Dana, Irvine, CA 92713. Booth #405; INFO/CARD #158.

The Polarad SMPD synthesized signal generator covers the frequency range of 5 kHz to 1 GHz with a resolution of 1 Hz.



#### Polarad Electronics

The output level range is +13 to -143 dBm with a resolution of 0.1 dB. Amplitude, Frequency, phase, and pulse may be modulated. Linear and logarithmic frequency

sweeps are available, with both the sweep width and sweep time being freely selectable. THe SSB phase noise is <-143 dBc/Hz at 20 kHz from the carrier (at 100 MHz). Polarad Electronics, Lake Success, NY 11042. Booth #311; Info/card #177.

Option 07 gives the Tektronic 496 and 496P portable spectrum analyzers 75



Tektronics

ohm measurement capability in addition to the standard 50 ohm measurement. Calibration is provided so that it is no longer necessary for the user to convert the units or account for external losses when changing from 50 ohm to 75 ohm. Option 07's frequency range of 5 MHz to 1 GHz, and 300 kHz resolution provide measurement capability for all cable and broadcast systems. The 496/496P offer spectrum analysis and measurements in the 1 kHz to 1.8 GHz range with 80 dB dynamic range. Tektronix Inc., Beaverton, OR 97077. Booth #504; please circle INFO/CARD #157.

The HPIL (Hewlett-Packard Interface Loop) available as an option on the Texscan SSG-1000 and SSG-2000 signal generators enable a low-cost hand-held computer, such as the HP-41C, to control the instrument. The SSG-1000 covers the frequency range of 100 kHz to 1000 MHz with a resolution of 10 Hz. The SSG-2000 extends the range to 2000 MHz, with a resolution of 20 Hz above 1000 MHz. Spurious noise is below -60 dBc (-54 dBc above 100 MHz), and harmonics are below -30 dBc. AM, FM, and PM are available. Texscan Instruments Corp., Indianapolis, IN 46226. Booth #417; please circle INFO/CARD #156.

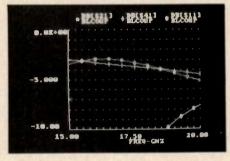
#### **RF** Software

The growing popularity of the IBM PC led Compact Software to develop a version of SUPER-COMPACT for the PC. SUPER-COMPACT PC models circuits as a series of two-ports. Up to fifteen variables may be optimized simultaneously. An important capability is the easy interface to the mainframe versions of the program. The initial analysis can be performed on the PC, and then uploaded to a faster mainframe for the "numbercrunching." Compact Software. Booth

#### #100, 101; INFO/CARD #155.

CADEC from Communications Consulting allows circuit analysis and optimization in the linear frequency domain. The circuit, described using the nodal method, is converted internally to a series of twoports. The program includes a fast Fourier transformation of calculated frequency points to a large variety of user defined waveforms. CADEC is available for IBM PCs and compatibles and the Hewlett-Packard Series 200. Communications Consulting Corp., Upper Saddle River, NJ 07458. Booth #515; please circle IN-FO/CARD #154.

Touchstone/RF is a version of EEsof's Touchstone program that has been specially tailored for RF work. The costly microwave circuit models such as microstrips and striplines have been eliminated, reducing the cost by about a half. Touchstone/RF is available for the HP Series 200 computers and the IBM PCs and



EEsof

compatibles. The program is compatible with their MICAD circuit layout program. EEsof, Inc., Westlake Village, CA 91362. Booth #610; INFO/CARD #153.

#### **EMC/EMI**

Compared to ordinary adhesive-bonded gaskets, Chomeric's vulcanized assemblies provide the following advantages: improved reliability, uniform sealing, simpler installation, and easier maintenance. Conductive, non-conductive, and non-conductive/conductive Combo elastomer gaskets can be vulcanized or bonded directly onto metal covers and enclosure panels. Chomerics Shielding Technology, Woburn, MA 01888. Booth #318; INFO/CARD #152.

Keene Corp., Ray Proof Division offers RF shielded rooms for secure communications, RF secure conference rooms, secure computer rooms, RF rooms for TEMPEST and testing, anechoic chamber systems, microwave/millimeter wave absorbers, EMI power line filters, and special purpose filters. Keene Corp., Ray Proof Division, Norwalk, CT 06856. Booth #401; INFO/CARD #151.



### For various counts of fast delivery, operating reliably, exceptional performance, and taking a tough stand on quality.

#### **AMPLIFIERS**

Sneaky critters. Don't make hardly any noise. Use a variety of disguises: flatpack, relay header, TO-8, even connectorized boxes. Operate all over the range—from 0.5 to 5200 MHz. Known to hang out with the notorious LOG AMP, a non-linear operator with a reputation for mighty wide dynamic range—some say 80 dB!

#### **CONTROL DEVICES**

a.k.a. (also known as) SWITCHES trigger-happy showoffs, some as fast as 50 ns!, MODULATORS—sharpies that're really up on the Code of the West, ATTENUATORS—tough guys that'll step you right down, and, watch out for these guys, PHASE SHIFTERS!

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More socially acceptable, double-balanced and all, but tough customers nonetheless. TIM,\* a particularly ornery type, will go with anything—doesn't care how he's loaded. Very much home on the range from a lowdown 0.02 MHz way up to 18 GHz! That's microwave, pardner.

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Alias for a powerful bunch of multiple personalities that will do anything for a buck. Their specialty is custom jobs. And they've got the most advanced tools and techniques of the trade. Fearless and experienced, they've even stood up to MIL-STD 883B!



#### PASSIVES

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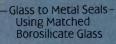
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> Alumina Substrate-**Expansion Coefficient** Matched to Header



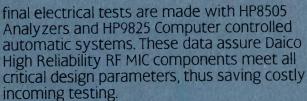


Actual size

# High Reliability RF MICS

### **DAICO High Reliability RF Microwave Integrated Circuit Components**

are products of state-of-the-art thin-film technology. Quality screened components, manufactured and tested to rigid Daico standards and numerous in-process quality tests, assures each device meets applicable MIL STD Specifications. Quantitative 100%





• Trun-film MIC construction. • DC-4GHz SPST-SP24T Speeds < 5nS</li>

10-90 RF1 Internal TTL or CMOS drivers

Internal 50 ohm terminations

PC Board mount Stripline or Connectorized



MIC VOLTAGE CONTROLLED ATTENUATORS Thin film MIC

- construction Linear attenu-
- ation to 60dB Linearity to
- 0.75dB
- Internal drivers Constant 50 ohm
- impedance
- PC Board mount Stripline or Connectorized

# **MIC STEP**

**ATTENUATORS**  Thin-film MIC construction

- Steps 1
- through 7 Attenuation
- to 63.5dB Speeds < 5nS</li>
- (10-90 BF Internal TTL or
- CMOS drivers
- PC Board mount or
- Connectorized



#### DAICO INDUSTRIES, INC.

2351 East Del Amo Blvd. Compton, Culif. 90220 Telephone: (213) 631-1143 · TWX 910-346-6741

INFO/CARD 30

# ONOLITHIC OROUADAU ICROVAU ICROVAU ICROVAU ICROVAU

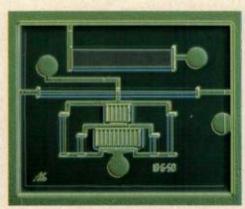
# Now! MMICs out of the lab and onto the shelf.

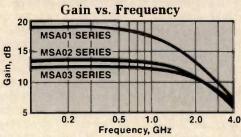
Avantek<sup>®</sup> MODAMP<sup>™</sup> silicon monolithic microwave amplifiers are the new alternative to hybrid modules—small, highly-reliable cascadable building blocks for frequencies through 3 GHz. And, they are now in production and available from stock.

Available in micro-X and 70mil transistor packages, MODAMP monolithic amplifiers permit greatly increased circuit and packaging density, while meeting the performance and reliability requirements of both military and commercial applications.

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MODAMP silicon monolithic amplifiers are now available, and priced from \$5.00 to





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# **Avantek**

Semiconductor Marketing 3175 Bowers Avenue Santa Clara, California 95051 (408) 727-0700

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# The SAW Resonator: How It Works

Part I gives an introduction to the Surface Acoustic Wave resonator and a TI-59 program to model device responses.

By Jeff Schoenwald Contributing Editor

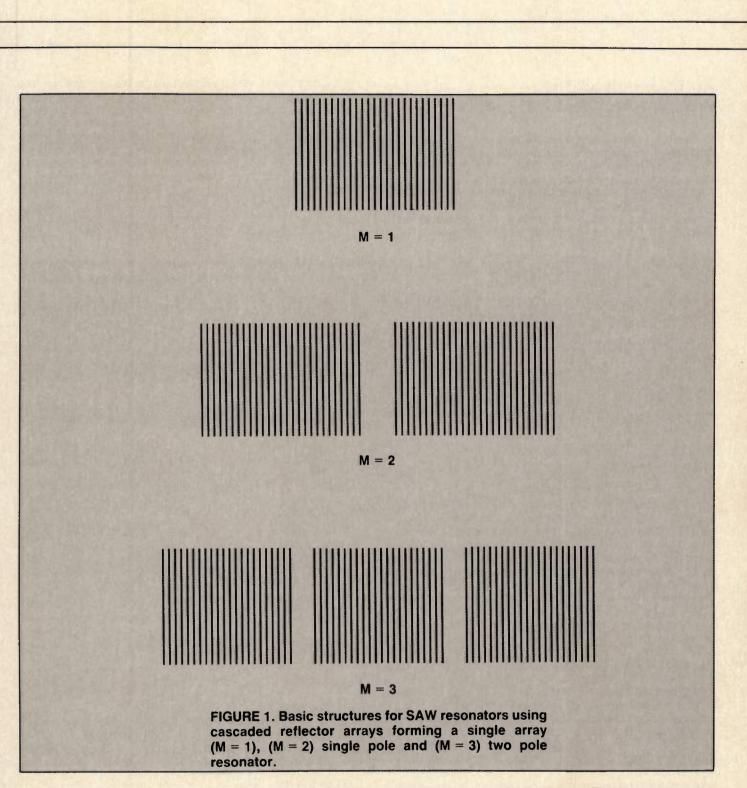
Long-time readers of *RF Design* have encountered the Surface Acoustic Wave (SAW) resonator before<sup>1,2</sup>. The original concept belongs to Eric Ash, of the Imperial College in London, who made the first attempt to produce a high Q resonant acoustic device that is planar in its fabrication (like all semiconductor circuits today), uses SAW, and can scale the frequency spectrum well beyond the practical limits achievable, then or now, by bulk quartz crystal filters. The first truly successful devices were demonstrated by Texas Instruments, followed soon after by scores of industrial, government and university laboratories in the United States, Europe and Japan. The theory, fabrication and performance of the SAW resonator is well chronicled in some detail and variation in the proceedings of the IEEE Ultrasonics Symposium going to back to 1974.

In this article I have tried to provide the RF engineer with a useful tool — a model with which he can simulate the many different device responses the SAW resonator has been found able to produce. In Part I the theory of the model is presented, along with a listing for the TI-59 calculator equipped with a printer. In Part II we manipulate the model to produce various devices and develop a good feeling for how performance depends on design and material parameters.

The first SAW resonator was a single pole device — a pair of reflecting structures forming a single cavity. Each reflecting structure consists of a repetitive set of lines and gaps — a grating. The grating is formed in various ways but always involves the use of vacuum deposition of metal (mostly aluminum) or occasionally a dielectric, and photolithographic exposure, developing and etching. Sometimes the stripes of the reflector are patterned in etched aluminum; the more reliable technique used today is one in which grooves are plasma etched directly out of

the quartz substrate. Each line and groove is one quarter wavelength long in the direction of propagation of the wave, and is typically 50-200 wavelengths wide, at the designated center frequency. Each pair of line and groove has a very low coefficient of reflection, purposely: too large a coefficient and the surface wave would quickly scatter into the bulk and dissipate. A distributed reflector consisting of several hundreds of lines (sometimes more than a thousand) produces a very efficient phase matched coupling between a SAW wave traveling in one direction and a reflected wave. As we shall see in some of the examples in Part II, the reflectivity can be very nearly 100%. Two such reflectors can form a cavity whose length is an integer number of half wavelengths. This is a Fabry-Perot cavity, and is the fundamental structure known as the single pole resonator. Place an interdigital transducer in the cavity (at the proper position, of course) and we have a way to dump electrical energy into this acoustic cavity through the piezoelectric coupling coefficient of the substrate. Remember that crystal guartz, lithium niobate, zinc oxide and many other materials are piezoelectric. Read my two previous articles<sup>1,2</sup> for a bit more background, or plow through the rich literature on the subject if you have a few years.

We needn't stop with a single pole resonator. Refer to Fig. 1. A third reflective array properly placed in line with the other two forms a second cavity — the two-pole SAW resonator. This may be continued indefinitely. The calculator is dumb and will not mind. But we, as engineers, can play the program included here like an arcade game and produce interesting and varied results. First, I'll describe the theory behind the program. Second, familiarize yourself with initial data entry procedures and device design. Then, next month, we'll start to build a succession of



devices and look at the power transmission and reflection characteristics of each one and, for a few selected cases, the change in performance of a single design as we make small changes in one of the parameters.

The program comes in two versions. The first one produces a listing of transmitted and reflected power as a function of frequency. The second version, which is run by replacing the last part of the program, will list the phase of the transmitted and reflected waves versus frequency. Unfortunately the TI-59 does not have enough memory to compute both and still be "user friendly" enough to handle all the I/O commands for data entry and listing on the printer. As a serious TI-59 programmer, you could generate more available memory by loading all the data off-line, eliminating the print commands to list them (jot them down elsewhere), and leave only the appropriate RUN/STOP commands to read important computed parameters as the program generates them at the beginning — you'll understand what I mean when we cover data entry. If you do all this, there should be enough memory space left to compute magnitude *and* phase of transmitted and reflected waves. I have chosen instead to rely on the automation features available on the TI-59 when used with the printer.

#### An Overview of the Program Architecture

After initial data entry, the program accomplishes its task by transmission line matrix computation methods. The 2x2 transmission matrix (T[1]) of a single line and gap segment is computed. The eigen values of this matrix are determined and the matrix T[N] of an array of N such segments is computed using Sylvester's Theorem (3). This theorem states that if T is a jxj matrix (in this case 2x2), then any polynomial matrix expression in powers of T may be written as:

$$P(T) = \sum_{r=1}^{J} \left\{ \begin{array}{c} P(\lambda r) & \prod_{s=1}^{J} \left( \frac{T - \lambda_s I}{\lambda_r - \lambda_s} \right) \right\}, I = \begin{bmatrix} I & O \\ O & I \end{bmatrix}$$
(1)

where I is the 2x2 identity matrix,  $\lambda r$  is the r'th eigenvalue of the matrix T, and P( $\lambda r$ ) is the same polynomial as before, but in powers of  $\lambda r$ , instead of the matrix T. For our 2x2 situation, j=2, T is the matrix of one section (a line and a gap), and P(T)=T<sub>N</sub>

Then:

$$T^{N} = \lambda_{1}^{N} \left( \frac{T - \lambda_{2} I}{\lambda_{1} - \lambda_{2}} \right) + \lambda_{2}^{N} \left( \frac{T - \lambda_{1} I}{\lambda_{2} - \lambda_{1}} \right)$$
(2)

$$= T \left( \frac{\lambda_1^N - \lambda_2^N}{\lambda_1 - \lambda_2} \right) + I \left( \frac{\lambda_1 \lambda_2^N - \lambda_1 \lambda_1^N}{\lambda_1 - \lambda_2} \right)$$

The eigenvalues  $\lambda$ , and  $\lambda_2$  are found by setting the determinant of T-l equal to zero and solving Eq. 2. This approach enables the closed-form, single pass calculation which is independent of the number of sections N in the grating array. The original version of this calculation did not use Sylvester's Theorem and relied instead on a recursive calculation which consumed time proportional to the size of the grating number N(4).

The matrix T[g] representing the gap of length  $I_g$  (normalized relative to the wavelength at the specified reference frequency  $f_0$ ) is computed and the structure cascaded by multiplication of the two matrices

$$T[N] \times T[g] = T[N+g]$$
(3)

The composite matrix is then solved for its eigenvalues and Sylvester's Theorem is used again to find the net transmission matrix of M such sections, which form M-1 cavities, or an (M-1) — pole resonator:

$$(T[N+g])^{M} = T[M]$$
<sup>(4)</sup>

Again, we must compromise because of the calculator's capacity. A practical device could have several cavities with different values of  $I_g$ , and the grating arrays between these cavities, referred to as coupling relfectors, might have different sizes. Nevertheless, the symmetric devices we are able to design are instructive, and represent the great majority of designs that have been practically implemented over the past 10 years.

The program assumes the transmission line structure is lossless, i.e., the impedance values are not complex. Making provisions for a complex impedance requires an ability to manipulate them in matrix multiplication, which the TI-59 does not do directly, or doubling the dimension of matrices and increasing the size Data entry procedures are summarized in the flow diagram in Table 1. After 11, 12, n2 and f0 are entered, fc, the "center frequency" at which the first maximum in reflectivity occurs, is computed:

$$fc = \frac{f0}{2(11+n2\bullet12)}$$
(5)

Other reflectivity maxima occur at odd harmonics of fc. The fractional "phase length" of a strip relative to a strip gap pair is

$$= \frac{n2 \cdot l2}{l1 + n2 \cdot l2}$$
(6)

If t = 0.5,  $|1 = n2^*|2$  and each segment has equal "optical," or phase thickness. The fractional difference in phase thickness between a gap and a strip is 1-2•t.

Armed with this input and a starting frequency, we may proceed to compute the transmission matrix of a strip-gap pair:

$$T[1] = \begin{bmatrix} A1 & B1 \\ C1 & D1 \end{bmatrix}$$
(7)

where

A1= 
$$\frac{(Z+1)\cos\gamma}{2} - \frac{(Z-1)\cos(1-2t)\gamma}{2}$$
 (7a)

B1=j 
$$\left(\frac{(Z+1)\sin\gamma}{2} - \frac{(Z-1)\sin(1-2t)\gamma}{2}\right)$$
 (7b)

$$\frac{C1=j}{2Z}\left(\frac{(Z+1)\sin\gamma}{2Z}+\frac{(Z-1)\sin(1-2t)\gamma}{2Z}\right)$$
(7c)

$$D1 = \frac{(Z+1)\cos y + (Z-1)\cos(1-2t)y}{2Z}$$
(7d)

and

γ

Once T[1] is obtained, we must find its eigenvalues in order to compute T[1]  $\land$  N = T[N], the equivalent of cascading N sections. We define the quantity 0,

$$\frac{\cos \Theta}{2} = \frac{A1+D1}{2}$$

$$\lambda_{1} = \cos \Theta + \sqrt{\cos^{2}\Theta} - 1^{2}$$

$$\lambda_{2} = 1/\lambda_{1}$$
(9)

If  $\cos^2\Theta - 1$  is zero or positive, the eigenvalues of t[1] are real. However, it is entirely possible that (A1 + D1)/2 < 0, in which case the eigenvalues are a complex conjugate pair. The distinction is made because the program branches to handle each case differently. If the eigenvalues are real, we obtain:

$$[\mathsf{TN}] = \begin{bmatrix} \mathsf{AN} & \mathsf{BN} \\ \mathsf{CN} & \mathsf{DN} \end{bmatrix}$$

where

$$AN = A_1 \frac{\lambda_1^N - \lambda_2^N}{\lambda_1 - \lambda_2} - \frac{\lambda_1^{N-1} - \lambda_2^{N-1}}{\lambda_1 - \lambda_2}$$
(11a)

(10)

(13)

$$BN = jB_1 \frac{\lambda_1^N - \lambda_2^N}{\lambda_1 - \lambda_2}$$
(11b)

 $CN = jC_1 \frac{\lambda_1^N - \lambda_2^N}{\lambda_1 - \lambda_2}$ (11c)

$$\frac{\mathsf{DN}=}{\lambda_1-\lambda_2} \frac{\lambda_1^{\mathsf{N}}-\lambda_2^{\mathsf{N}}}{\lambda_1-\lambda_2} - \frac{\lambda_1^{\mathsf{N}-2}-\lambda_2^{\mathsf{N}-1}}{\lambda_1-\lambda_2}$$
(11d)

If the eigenvalues are complex, we have instead

AN=	D sin N $\Theta$ _ sin (N-1) $\Theta$	
	sin O sin O	(12a)
BN=	jB1 sin NO	
	sin Ø	(12b)
CN=	jC1 sin NO	
	sin Θ	(12c)
DN=	D1' $\sin N\Theta - \sin (N-1)\Theta$	
	sin Θ sin Θ	(12d)

where  $\Theta$  is obtained from (9).

Having constructed a grating array [TN], we must now compute the transmission line that will make up the cavity between two reflectors. The matrix for this section is simple:

where

$$\gamma = 2\pi \left( \frac{f_1}{f_c} \quad lg \right)$$
(14)

The matrix for the cascaded reflector and gap is simply obtained using (1). The same calculation is applied to T[N+g] to find its eigenvalues, and then the composite structure is cascaded M times to build a transmission line with M reflectors and M-1 cavities. Note that the tail end of the structure contains a length of transmission line  $I_g$ . This will have no effect on the

Calculate |T (1) | at f<sub>1</sub>  
Eigenvalues of |T (1)|  
Cascade array + gap  

$$|T(N + 9)|$$
  
Eigenvalues of |T (N + 9)|  
Print f<sub>1</sub>, reflection, transmission  
Increment f<sub>1</sub> + df  $\rightarrow$  f<sub>1</sub>  
No  
Is f<sub>1</sub> > f<sub>2</sub>?  
Yes  
STOP

Program Structure Data Entry (I<sub>1</sub>, I<sub>2</sub>, n<sub>2</sub>, f<sub>0</sub>, Z, f<sub>1</sub>, df, f<sub>2</sub>, M, Ig)

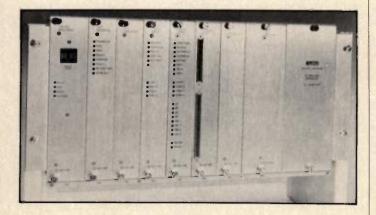
TABLE 1

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magnitude of the power transmitted or reflected, but only a phase shift in the transmitted wave.

Once the transmission of matrix T of the entire structure has been computed, the normalized transmission and reflection amplitudes Tf and Rf are easily computed:

Tf=	2 A'+B'+C'+D'	(15a)
Rf=	A'+B'-C'-D' A'+B'+C'+D'	(15b)

Since the device is assumed to be lossless only one coefficient need be calculated, say Tf, and

RF 2=1- Tf 2.

### (16)

### Using the Program for Design and Analysis

The program is listed in Table II. Partition the TI-59 for 3op17 (719.29) and load the program. Two magnetic cards are needed to store it. Enter the data as shown in Table III. Each input parameter is printed as it is entered. After  $f_o$  is entered and printed,  $f_c$  is computed and it too is printed. This is done to aid the user in choosing an appropriate frequency range for analysis. After all input data has been entered, calculation commences, resulting in a listing of the frequency and corresponding normalized reflected and transmitted power (in dB). A complete cycle takes about 20 seconds. The computation is then repeated at the next frequency until the entire frequency domain has been examined. While the program chugs away, take a break.

The user may FIX the number of decimal places desired at the beginning or at any point during data entry, since, for example, the reflection or transmission loss may only be interesting to 0.1 dB accuracy. Care must be taken, however, since the printed listing of input parameters or frequency may appear truncated to less accuracy than was originally specified.

Join us next month for Part II. We will use the program to examine the properties of reflective arrays, single and multiple cavities. As an added bonus, a BASIC version is also provided for the growing number of microcomputer users.

### References

1. J. Schoenwald, "Surface Acoustic Waves for the R.F. Design Engineer," *r.f. design*, pp. 11-16, March/April 1981.

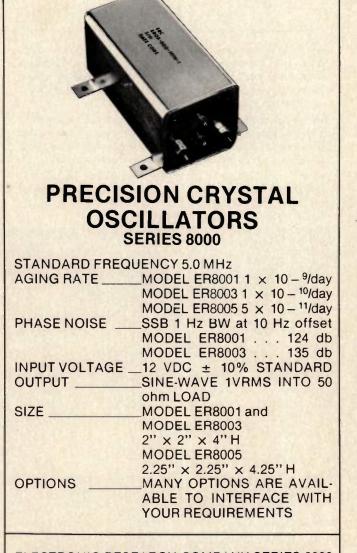
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 J. Irving and N. Mullineau, Mathematics In Physics and Engineering, pp. 285-288, Academic Press, New York, N.Y. (1959).
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 E.K. Sittig and C.A. Caquin, "Fitters and Dispersive Delay Lines Using Repetitively Mismatched Ultrasonic Transmission Lines," IEEE Transactions on Sonics and Ultrasonics, Vol. SU-15, pp. 111-119, (1968).

	TABLE III	
	Data Entry	
Parameter	Enter	Press
gap length	4	Α
line length	12	R/S
veolocity index	Π2	R/S
reference frequency	fo	R/S
impedance ratio	Z	В
number of sections	N	R/S
start frequency	f <sub>1</sub>	С
frequency increment	df	R/S
stop frequency	f2	R/S
number of relfectors	M	R/S
cavity size	lg	R/S

1						
		Ta	able II			
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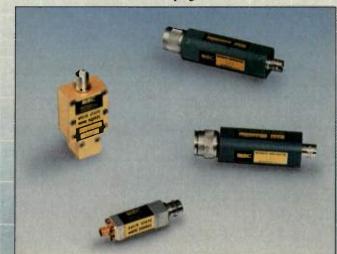
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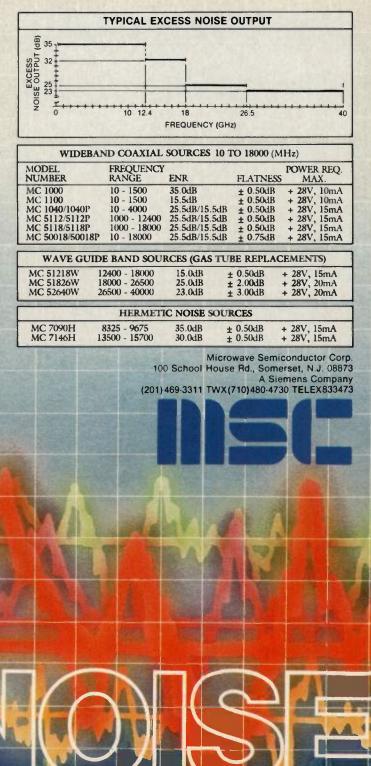
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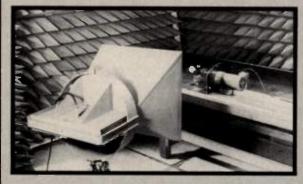
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Table II continued

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# The Phase/Frequency Detector

An analysis of phase-locked loop design employing phase frequency detectors.

#### By James Crawford

The controversial subject of "Divider Time Delay"<sup>1</sup> in recent *RF Design* issues has prompted the following discussion which was presented at a Hughes Aircraft Co. in-plant class on phase-locked loop design. The following analysis is an endorsement of Dr. Egan's explanation<sup>2</sup> where he states that the appearance of the delay-like term is due to the sampling process which is taking place in the phase-locked loop, not the transport delay or any other delay through the divider.

The delay-like terms which was mentioned above is shown in equations (3) and (4) of reference [3]. In this reference it is suggested that "the discrepancy between theory and experiment (in phase-locked loop design) was found to be attributed to divider delay which caused a decrease in phase margin significant enough in many cases to cause unstable loop performance." A decrease in loop phase margin does indeed occur in these phase-locked loops but the sole mechanism is a result of the sampling process which is taking place in the closed loop.

A rigorous analysis of phase-locked loop design employing phase-frequency detectors necessitates a detailed examination of the operation fundamentals. Rather than deal immediately with the specifics of phase-locked loop design using the phase/frequency detector, the problem will be dealt with using the following approach:

1) A general discussion of sampling phase-locked loops will be given which will display some of the differences between the true open-loop gain function, and the commonly used continuous approximation to the open-loop gain function.

2) With the sampling basics now developed, the transfer function for the phase/frequency detector will be found in some detail.

3) The phase/frequency detector transfer function is used to write an accurate impulse for the open-loop gain function. Given this function, a band-limited approximation of the open-loop gain function will be found using Z-transforms. The final band-limited expression can be used with conventional continuous transform (Laplace transforms) design methods for phase-locked loops.

### Sampling Phase-Locked Loop Fundamentals

In contrast to the continuous mixer-type phase detector, the phase/frequency detector is a sampling phase detector. Phase error information is available at discrete time intervals which are spaced at exact intervals of T seconds, where T is the period of the reference frequency. The phase error at each instant is in the form of an impulse function whose area is proportional to the phase error. In reality, the impulses out of a phase-frequency detector such as the Motorola 4344 are of finite amplitude, but this fact is completely negligible in light of the RC filter time constants which follow the device. The phase detector output pulses may be mathematically viewed as ideal

impulse functions of appropriate area. For the time being, the concept of the phase/frequency detector as an ideal impulse sampler will be deferred and developed momentarily.

The phase-frequency detector is modeled in Figure 1 as an ideal impulse sampler. The function H(s) represents some form of analog "hold" function such as the RC lowpass filters that customarily follow a phase-frequency detector. The function G(s) represents the normal loop filter transfer function.

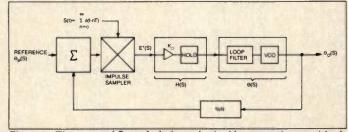


Figure 1. The general Sampled phase-locked loop employs an ideal impulse Sampler which must be followed by some form of "hold" device. (HS).

Before jumping into the loop details, as an aside, consider the time function f(t) which is sampled by an ideal impulse sampler. The sampled time function can be written as in (1) where the asterisk (\*) represents the time-sampled form of the function.

$$f^{*}(t) = f(t) \sum_{n=0}^{\infty} \delta(t - nT)$$
(1)

From Laplace platform and convolution theory, (1) may be written in Laplace transforms as in (2) where  $F^*(s)$  and  $f^*(t)$  are Laplace transform pairs. Note that F(s) and f(t) are also Laplace transform pairs.

$$F^{*}(s) = F(s) \star L \left\{ \sum_{n=0}^{\infty} \delta(t - nT) \right\}$$
(2)

The  $\star$  in (2) represents convolution in the frequency domain. Equation (1) may be used with the definition of the forward onesided Laplace transform to give yet another interpretation of the time-sampled function form in the frequency domain.

$$F^{*}(s) = \int_{0}^{\infty} f^{*}(t) \exp(-sT) dt$$
(3)  
$$= \int_{0}^{\infty} f(t) \sum_{n=0}^{\infty} d(t-nT) \exp(-sT) dt$$
$$F^{*}(s) = \sum_{n=0}^{\infty} \int_{0}^{\infty} d(t-nT) \exp(-sT) dt$$
(4)

60

46

$$F^*(s) = \sum_{n=0}^{\infty} f(nT) \exp(-sT).$$
(5)

Equation (5) is actually the defining relationship for the Ztransform of f(t). This fact will be used later in this article for easy calculation of  $F^*(s)$ .

A very powerful relationship may be found by continuing the convolution calculation in (2). Since the convolution must be performed in the frequency domain, we must know the Laplace transform of the infinite series of impulse functions which are performing the sampling operation. Since

$$\delta(t) \leftrightarrow 1 \tag{6}$$

then

$$L \{ \sum_{n=0}^{\infty} \delta(t-nT) = 1 + \exp(-sT) + \exp(-2sT) + \frac{1}{1 - \exp(-sT)}$$

The convolution of (2) may be rewritten as

$$F^{\star}(s) = F(s) \star \frac{1}{1 - \exp(-sT)}$$

$$= \int \frac{F(u) \, du}{1 - \exp(sT - uT)}$$
(8)

TZ

where 
$$Z = \exp(uT)$$
 and  $du = dZ/(ZT)$ 

1- exp(sT)/Z

Taking this process one step further and using the Residue theorem, this integral may be evaluated as in (9).

$$= \frac{1}{T} \int \frac{F(Z) dZ}{Z - exp(-sT)}$$

$$= \frac{1}{T} \sum_{n=-\infty}^{\infty} F(s+j n Ws) \qquad \text{where } Ws = 2 \pi/T.$$
(9)

Equation (9) is a valuable result and although it involves an infinite summation, it may be used to evaluate  $F^*(s)$ .

These previous transform tools will appear much more valuable if we now return to a discussion of Figure 1. As in classical PLL analysis, the most expedient first step in the loop analysis is to solve for the error function E\*(s). We may write

$$E^{*}(s) = (\Phi r - \Phi o)^{*}$$
  
=  $\Phi r^{*} - \Phi o^{*}$   
=  $\Phi r^{*} - [E^{*}(s) H(s) G(s)]^{*}$  (10)

Those unfamiliar with sampled systems will find [4] particularly useful and easy to understand. As developed in Chapter 4 of [4], the sampling operation in (10) may be brought within the brakcets because E(s) is already a sampled function. The sampled error function is then given by

$$E^{*}(s) = \Phi r^{*} - E^{*}(s) HG^{*}(s)$$
 (11)

We finally obtain the desired result for the sampled loop error function.

$$E^{*}(s) = \frac{\Sigma \Phi r(s+j n Ws)/T}{n = -\infty 1 + HG^{*}(s)}$$
(12)

RF Design

In most cases,  $\Phi r(s)$  can be assumed to be effectively bandlimited and aliasing of noise products can be neglected. This gives some simplication to (12) as given in (13).

$$E^{*}(s) = \frac{\Phi r(s)/T}{1 + HG^{*}(s)}$$
(13)

where  $\Phi^*(s) \sim \Phi r(s)/T$ .

(7)

The asterisks would be absent in classical analysis of a phaselocked loop which neglected sampling effects. With rare exception, most systems which use the phase/frequency detector have a small bandwidth compared to the reference frequency in order to obtain reasonably low "sampling spurs." For this reason, the higher order terms (terms other than n=0) in equation (9) can largely be ignored for low bandwidth situations. This is precisely why classical analysis ignoring sampling effects still provides excellent results in small bandwidth situations. As the loop bandwidth is increased compared to the reference frequency, however, the higher order terms cannot be ignored.

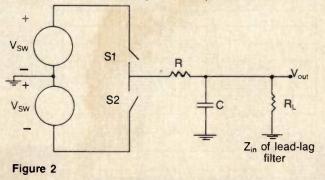
For large loop bandwidth situations (bandwidth > 0.1  $F_{ret}$ ), Ztransform techniques should be used to include the higher order effects. If there are true transport time delays within the loop, modified Z-transforms should be used. If the loop bandwidths remain small (say < 0.1  $F_{ret}$ ), bandlimited forms of the openloop gain function may be found which very accurately describe sampling effects without resorting to Z-transform analysis. Equation (9) which is repeated below as equation (14) will provide the menu for arriving at a bandlimited form of the open-loop gain function including sampling effects. The continuous Laplace transform impulse response, Gol(s), must first be found. The continuous open-loop gain function will be re-expressed in terms of Z-transforms and the assumption of small loop bandwidth imposed. The final result will be the bandlimited form of the openloop gain function with first order sampling effects.

$$F(Z) = \frac{1}{T} \sum_{n=-\infty}^{\infty} F(s+j n Ws)$$
(14)

It will be shown that the so called "divider time delay" appears during this step and is solely a result of sampling.

# Impulse Response of the Phase/Frequency Detector

A simplistic equivalent circuit of the phase/frequency detector is provided in Figure 2. No attempt has been made here to describe the frequency discriminator mode of operation. In many applications, the phase detector remains in its linear range of operation, because although the phase error may be very large at the VCO, it is reduced by N at the phase detector.



In steady-state operation, the switches in Figure 2 are open 95 percent of each reference period, only closing long enough to replenish the small discharge in capacitor C each reference period. (Since the 4344 type phase detector cannot resolve absolute time difference between the divider and reference pulse trains less than its own internal time delay, some built in offset is needed to avoid the detector's "dead zone" of operation at zero time difference between the two waveform trains.) During the period of frequency acquisition, the proper polarity switch is closed for a length of time which is defined by the time difference between the leading edge of the divide-by-N signal. This signal relationship is shown in Figure 3. The phase detector output pulse width is directly proportional to the phase error within the loop.

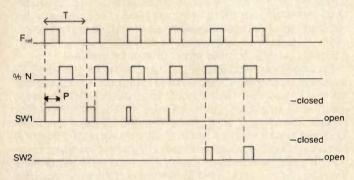
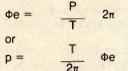


Figure 3

Pack cubit (1.7)



The pulse widths out of the phase detector, p, are very small with respect to the reference period because a Type II loop is always used (zero steady-state phase error) and the VCO phase error is reduced by the divider ratio, N.

We are primarily interested in the impulse response of the phase detector/lowpass filter combination. The transfer function for the lowpass filter alone is given by (16).

$$FL(s) = \frac{R1}{R1 + R} \frac{1}{1 + s \tau 1}$$
  
where  $\tau 1 = \frac{R1 R C}{R1 + R C}$ 

R1 + R

A typical pulse response of the circuit in Figure 2 is presented in Figure 4. The rising edge is very linear because  $p \ll \tau 1$ . The output voltage,  $V_{out}$ , can be easily found from Figure 4. The output voltage during the next reference period is given by (17).

$$V_{out}(t) = V_{o} \exp(-t/\tau 1) + \frac{R1}{R1 + R} V_{sw}(1 - \exp(-t/\tau 1))$$
(17)

(15)

(16)

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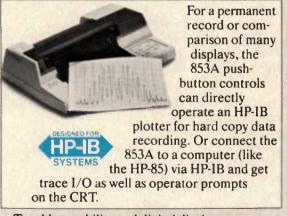
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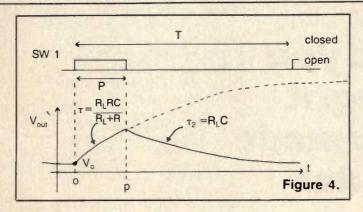
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#### for 0 < t < =p

 $V_{out}(t) = V_{out}(p) \exp(-t/\tau 2)$  for p < t < T.

In a properly designed Type II loop under normal linear operation,  $p \ll \tau 1$  and equation (17) may be written in a more simple form without the first exponentials.

$$V_{out}(t) \sim V_o + \frac{R1}{R1 + R} V_{sw}(t/\tau 1)$$
  
for 0 < t < = p

 $V_{out}(t) = V_{out}(p) \exp(-t/\tau 2)$  for p < t < =T.

It is important to note once more that  $V_0$  represents the initial capacitor voltage due to any previous phase sample,  $\Phi e$ .

The Laplace transform transfer function can be found by using the above equations directly.

$$F_{out}(s) = L\{(V_o + \frac{R1}{R1 + R} V_{sw}, \frac{p}{\tau 1}) \exp(-t/\tau 2)\}$$
  
(19)

 $\frac{V_{o}}{s + 1/\tau 1} + \frac{R1 V_{sw} \Phi e T}{(R1+R) \tau 1 2 \pi (s+1/\tau 2)}$ 

where the phase error is  $\Phi e = 2\pi p/T$ .

As stated earlier,  $V_o$  is a direct result of earlier samples of the phase error,  $\Phi e$ . Using this fact, it is possible to show that the output voltage as a function of the input phase error is given by (20).

$$\frac{V_{out(s)}}{\Phi e(s)} = \frac{Kd T}{(R1+R)/R1+sRC} \frac{1}{1-exp(-sT) exp(-T/\tau 2)}$$
(20)

where Kd =  $V_{sw}/(2\pi)$ 

This is the final result for the phase detector impulse response. The factor T is a direct result of the sampling operation. Notice that if  $T/\tau 2$  is not large, the second multiplicative factor cannot be ignored. In this case, the output voltage is a function of the present phase detector error as well as the previous error samples and Z-transform analysis is required. Since the intent of this analysis has been to eventually arrive at a band-limited form of the open-loop gain function, in that vein,  $T/\tau 2$  will be assumed to be >> 1 such that Z-transform analysis will not be required in the final end result.

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### Derivation of the Continuous Band-Limited Open-Loop Gain Function with Sampling

The phase-frequency detector is always used in a Type II phase-locked loop in order to realize reasonable spurious performance and tuning range. In order to simplify the mathematics involved, however, an example using a Type I system will be used. Our approach will be to calculate Gol(Z) and compare it to the continuous form of Gol(s). This will reveal the effects of sampling upon the otherwise continuous open-loop gain expression.

In order to make any connection between an impulse sampled system and a continuous system, the sampled loop must have some form of "hold" device which effectively converts the phase detector impulse functions into smooth time waveforms which have a finite width in time, and a finite height. If the "hold device" is not present, the loop *must* be analyzed as a sampled system. No equivalent continuous system would exist for that case.

The "hold" device may be as simple as an RC lowpass filter, or as complicated as a true 0-order sample/hold. Consider a continuous Type I phase-locked loop with a low pass "hold" as given in (21).

(21)

(22)

(23)

(24)

$$Gol(s) = \frac{Kd Kv}{N s} \rightarrow \frac{Wn}{s} \frac{1}{1+s}$$

where

Kd=phase detector gain Kv=VCO sensitivity N =feedback divider ratio

 $\tau$  =low pass filter time constant representing the "hold."

As shown earlier, the sampling effects upon a continuous function may be included by taking the Z-transform of the time function provided that the continuous function is correct of course. In the previous section, it was shown that the impulse response of the phase/frequency detector followed by a simple RC lowpass filter is given by (20). Therefore, assuming that  $T >> \tau 2$  in equation (20), equation (21) must be multiplied by T to have proper form. Using a table of Z-transforms, equation (21) may be easily converted into Gol(Z).

 $\frac{\text{Wn T}}{\text{s(1+s\tau)}} \rightarrow \text{Wn (1- exp(-t/\tau)) T}$ 

$$\frac{\text{Wn Z T}}{7-1} - \frac{\text{Wn Z T}}{7-4}$$

Collecting terms in (22), the Z-transform for the Type I system is simply given by (23).

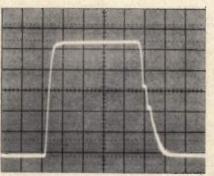
Gol(Z) = 
$$\frac{\text{Wn Z (1-A) T}}{(Z-1)(Z-A)}$$

We can effectively remove any significant effects of the "hold" device upon the system by allowing  $\tau \rightarrow 0$ , i.e.,  $A \rightarrow 0$ . This is equivalent to making the RC filter time constant negligible compared to the reference period, T. In the limit as  $A \rightarrow \Phi$ ,

Gol(Z) = 
$$\frac{\text{Wn Z T}}{(Z-1)(Z-0)} = \frac{\text{Wn T}}{(Z-1)}$$

Equation (24) may be expressed in terms of the more familiar complex frequency, s, by noting that Z = exp(sT). Doing so, we obtain (25).

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Number of harmonic terms included in summation 5 Loop reference frequency 10000 Typ 1 loope Wn 2000 Loop LPF time constant, nsec 5000

APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION

				10003 001
	Summation o	f Gol Terms	Exp(	-57/2)
F	Gol,dB	Ang,Deg		
100	10.06	-91.28	18.86	-91.98
158	6.53	-91.88	6.54	-92.97
288	4.83	-92.40	4.84	-93.96
388	8.51	-93.61	0.51	-95.94
488	-2.88	-94.81	-1.98	-97.92
500	-3.94	-96.03	-3.92	-99.98
788	-6.89	-98.47	-6.85	-183.86
1000	-18.83	-182.21	-9.95	-109.00
1508	-13.65	-188.72	-13.47	-119.78
2888	-16.28	-115.74	-15.98	-129.68
3666	-20.07	-132.17	-19.52	-149.38
4886	-22.55	-153.18	-22.05	-169.16

...........

Number of harmonic terms included in summation 10

Loop reference frequency 10000

Type I loop Wn 2000 Loop LPF time constatn, nsec 5000

APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION

	Summation o	f Gol Terms		-ST/2)
F	Gol,dB	Ang, Deg		
100	10.06	-91.46	18.86	-91.98
158	6.54	-92.19	6.54	-92.97
200	4.84	-92.93	4.84	-93.96
388	0.51	-94.39	8.51	-95.94
400	-1.98	-95.86	-1.98	-97.92
588	-3.92	-97.33	-3.92	-99.90
788	-6.84	-188.38	-6.85	-183.86
1898	-9.94	-104.79	-9.95	-189.80
1588	-13.45	-112.49	-13.47	-119.70
2888	-15.92	-128.54	-15.98	-129.68
3000	-19.24	-138.14	-19.52	-149.38
4999	-21 15	-158 88	-22 85	-169.16

Number of harmonic terms included in summantion 5

Loop reference frequency 1000

Type II loop Wn 2000

Loop damping factor, eta .707 APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION

	Summation of	Gol Terms	Continuous G Exp( -ST/2	
F	Gol,dB	Ang, Deg		
160	20.95	-156.41	28.98	-158.83
150	14.77	-147.09	14.67	-149.29
200	10.75	-139.68	18.68	-142.34
300	5.68	-129.45	5.46	-132.82
468	2.48	-123.29	2.22	-127.29
568	8.16	-119.51	-0.11	-124.14
700	-3.13	-115.77	-3.41	-121.69
1000	-6.45	-114.59	-6.72	-122.49
1500	-10.16	-117.17	-18.37	-128.23
2000	-12.79	-122.18	-12.92	-136.02
3666	-16.58	-136.40	-16.49	-153.68
4866	-18.84	-155.59	-19.03	-172.39

Number of harmonic terms included in summation 10 Loop reference frequency 1000

Type II loop Wn 1000

Loop LPF time constant, nsec 5000

APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION

	Summation or	f Gol Terms	Exp( -ST/2 )		
F	Gol,dB	Ang, Deg			
100	10.70	-139.13	10.60	-148.36	
150	5.60	-128.39	5.46	-129.85	
200	2.38	-121.69	2.22	-123.33	
368	-1.73	-114.49	-1.93	-116.51	
488	-4.46	-111.20	-4.67	-113.64	
588	-6.51	-109.71	-6.72	-112.59	
768	-9.53	-189.28	-9.75	-112.99	
1888	-12.68	-111.03	-12.98	-116.22	
1500	-16.21	-116.62	-16.46	-123.99	
2888	-18.67	-123.59	-18.98	-132.82	
3868	-21.97	-139.95	-22.53	-151.53	
4888	-23.85	-158.98	-25.06	-178.77	

$$Gol(s) = \frac{1}{i2} \frac{Wn \exp(-sT/2)T}{\sin(wT/2)}$$

For frequencies which are small compared to the reference frequency,  $F_{ref} = 1/T$ , the sin(x)  $\sim$  x approximation may be made, reducing (25) to finally (26). This is equivalent to the initial premise that the loop bandwidth is much less than the reference frequency.

$$Gol(s) =$$
 Wn exp(-sT/2)

S

The final result for the bandlimited form of the open-loop gain function is given in (27). This expression includes the first order sampling effects. The appearance of the so-called time delay exponential occurred without introducing any transport time delay whatsoever, only the sampling effects. (27)

$$Gol(s) = \frac{exp(-sT/2) Wn}{s}$$

Generalizing, first order sampling effects for phase-locked loops which have a small percentage bandwidth compared to the reference frequency can be analyzed using classical Laplace transform methods provided the new "delay term" is included in the phase detector transfer function.

### **Further Proof**

As further proof of our result above, we may compare this result with that obtained using (9). Only the first few terms of the infinite summation in (9) will be included. The computer program and sample run appear in Appendix I. Notice that the inclusion of the added exponential term of (28) with the normal Type I open-loop gain results in very good agreement between the two mathematical models for frequencies well within the closed-loop bandwidth. The phase of the open-loop gain function would be very inaccurate had the exponential term been left out. As the loop bandwidth increases with respect to the reference frequency, the approximation shows more and more deviation from the true open-loop gain calculated by (9). [Note that for all cases,  $T/\tau 1 >> 1$  has been assumed with loop bandwidth  $<< F_{ref}$ .]

In order to be complete, the same calculation was performed for the Type II phase-locked loop with a phase-frequency detector and small RC lowpass filter "hold." The continuous form of the open-loop gain function is given in (29) where T is due to the phase detector transfer function.

$$Gol(s) = \frac{NOT}{1+s\tau} \frac{1+s\tau 2}{s\tau 1} \frac{NV}{Ns}$$

Our approximation to G\*ol(s) is found using equation (28).

(20)

(25)

(26)

(28)

G\*ol(s)~ exp(-sT/2) 
$$\frac{\text{Kd}}{1+s\tau} \frac{1+s\tau^2}{s\tau 1 \text{ N}} \frac{\text{Kv}}{s}$$

The true function G\*ol(s) is found again from substituting equation (29) into (9). (31)

$$G^{*}ol(s) = \frac{1}{\Sigma} \sum_{\substack{T = -\infty \\ T = -\infty}}^{\infty} \frac{Kd T}{1+u\tau} \frac{1+u\tau^{2}}{u\tau^{1}} \frac{Kv}{Nu}$$

where u =s + jnWs

Reiterating, the T following Kd is due to the phase detector transfer function, (20), whereas the 1/T is due to the leading coefficient in (14).

A second computer program and sample run are provided in Appendix I for this Type II phase-locked loop case. Once again, the bandlimited gain expression in (30) closely approximates the true gain function (31) for frequencies well within the closed-loop bandwidth.

In concluding this article, several statements stand out.

• In order to make any correlation between sampled and continuous systems, some form of analog "hold" device impulse response of the hold device is the interpolating waveform between sample points in the time domain.

• The appearance of sampling in any loop with the accompanying "hold" device, causes an exponential phase term to appear, exp(-sT/2).

• Sampling effects in small percentage bandwidth loop (wrt.  $F_{ref}$ ) may be quite accurately described by the normal continuous open-loop gain function provided that the additional exponential term is included.

Sampling effects cause the appearance of the exponential delay-like term whenever quantities within the loop are only available at dicrete instants in time. Digital dividers within the feedback loop in an otherwise continuous loop will still cause the exponential term to appear. If digital feedback dividers and a digital phase/frequency detector are used within the phase-locked loop together, only one exp(-sT/2) term results (the continuous first order approximation remains unchanged). True

transport delay within the loop is accounted for by another exponential delay term. (Here, true transport delay refers to delay through op-amps and propagation delay refers to delays through other components, including dividers. These quantities are available in component data books.)

Analysis of the phase/frequency detector in a phase-locked loop can be considerably more complex than the usual continuous analysis which is generally employed. If the loop bandwidth is, say,  $< F_{rel}/40$ , the loop can be considered continuous for all practical purposes. For higher percentage bandwidths, attention should be given to the added "delay-like" term shown in equation (28) and care should be given to insure that  $T/\tau 2$ is > 3 in equation (20). [If  $T/\tau 2 < 3$ , another integrator in the form of a time variable filter is created which makes the situation much more complex. Of course, keeping  $T/\tau 2 > 3$  will result in higher spurs and notch filtering will undoubtably be required. The time variable filter increases the gain within the loop bandwidth and adds substantial phase as well which can easily lead to instability. For best results, choose  $T/\tau 2 > 3$ .]

Although the cautious aspects of sampled phase-locked loop design have been brought out for the phase/frequency detector, sampled systems harbor much more capability than first glance indicates. For instance, a Type I sampled loop which employs a zero order sample and hold rather than the phase/frequency detector will theorectically perform phase-lock in only one sample period! An ideal Type II phase-locked loop with a zero order sample and hold phase detector is capable of performing phase-lock in only two sample periods. These speedoptimized phase-locked loops must be analyzed using Z-transforms.<sup>5</sup>



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### References

1. RF Design, "PLL Primer, Part III," A.B. Przedpelski, July/ August 1983, pp. 48-58.

2. RF Design, Letters to the Editor, Dr. William F. Egan, March/ April 1984, pp. 9A-12A.

3. RF Design, "Divider Delay: The Missing PLL Analysis Ingredient," Stan Goldman, March/April 1984, pp. 58A-66A.

4. Digital Control of Dynamic Systems, Gene F. Franklin, J. David Powell, Addison-Wesley.

5. Microwaves & RF, "Sampling Phase-Locked Loops for Frequency Synthesis," J.A. Crawford (to be published).

### Appendix I

FILE: SAMPLE 71

PAGE BEI

28 38 48 Comparison of SumE F(s + jnHs) ] with 2-transforms

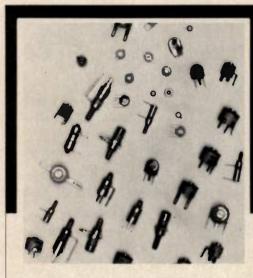
#### 58 \* DIM PTS(20) ,MAG(20) ,ANG(20) PI=3.1415926548

- 62
- 70 FOR IX=1 TO 17 80 READ PTS(IX) 90 NEXT IX

100 DATA 188, 158, 200, 300, 400, 500, 700, 1000, 1500, 2000, 3000, 4000, 5000, 7000 118 DATA 8000, 9000, 10000 128

- 130 INPUT 'NUMBER OF HARMONICS TO INCLUDE ', NHARGE. 135 INPUT 'INPUT REFERENCE RATE, H2 ', FREF 136 INPUT 'INPUT THE TYPE I LOOP WN ', WN

- 137 INPUT "INPUT THE LON-PASS FILTER TIMECONSTANT, NSEC" ;TAU 138 TAU=TAU19.999997E-18 158 FOR 1%=1 TO 17 168 SUMR=0 FOR J = 1 TO NHARD( F=PT5(1%) + J%\*FREF GOSUG 1000 SUMR=SUMR + FR 182 198 218 228 SUMI=SUMI + FI F=PTS(I%) - J%#FREF 238 235 GOSJ8 1888 SUMR=SUMR + FR SUMI=SUMI + FI 258 NEXT J% 268 278 F=FTS(12) GOSUB 1988 SUHT=SUHT + FR SUHT=SUH1 + F1 MAG(12)=4,342941LOG(SUHT^2 + SUH1^2) ANG(12)=4,342941LOG(SUHT^2 + SUH1^2) IF SUHT(0 THEN ANG(12)=MAG(12)=F1 SUHT(0 THEN ANG(12)=F1 SUH(0 THEN ANG(12)=F1 SUH(0 THEN ANG(12)=F1 SUH(0 THEN ANG(12)=F1 SUH(0 THEN ANG(12)=F1 SUHT(0 THEN ANG(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT(12)=F1 SUHT 288 308 328 ase IF SUMR(0 THEN ANG(1%)=ANG(1%)=PI 333 ANG(1%)=ANG(1%)II00/PI 340 HEXT 1% 350 PRINT CHR1(26) 351 UPRINT "Number of harmonic terms included in summation "(NHMPR). 352 UPRINT 353 LPRINT 'Loop reference frequency ';FREF 354 LPRINT 355 LPRINT 'Type 1 loop Wn ';WN 356 LPRINT \*Loop LPF time constant, nsec \*;TAUX1E+89 358 LPRINT 368 LPRINT \*APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION\* 378 LPRINT \* Continuous Gol 371 LPRINT \* Summation of Gol Terms Exp( -ST/2 )\* 388 LPRINT 'F Gol , dB Ano,Deo 488 FOR 1%=1 TO 17 401 F=PTS(1%) GOSUB 1000 NORGAIN=4.34294 # LOG( FR\*2 + FI\*2 ) 483 494 ......
  - NORANG=B WE .S. FREF ....... 418 ....... . . . PTS(1%) , MAG(1%) , ANG(1%) , NORGAIN , NORANGE 188/PI



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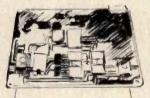
#### FILE: SAMPLE T2

PAGE ....

10 THERE ARE ARE ARE ARE ARE ARE ARE ARE ARE
20 '
30 ' Comparison of Suml F(s + jnWs) ] with Z-transforms
48 *
Se 'XRXARXXXXXXXXXXXXXXXXXXXXXXXXXXXXXXXXX
68 DIM PTS(28) ,MAG(28) ,ANG(28)
70 P1=3.141592654#
88 FOR 1%=1 TO 17
98 READ PTS(1%)
100 ND(T IX
110 DATA 100,150,200,300,400,500,700,1000,1500,2000,3000,4000,5000,7000
128 DATA 8000,9000,10008
130 '
148 INPUT "NUMBER OF HARMONICS TO INCLUDE " ; NHARMY.
150 INPUT "INPUT REFERENCE RATE, HZ ";FREF
160 INPUT "INPUT THE TYPE II LOOP IN "INN
170 INPUT "INPUT THE LOOP DAMPING FACTOR ";ETA 180 INPUT "INPUT THE LOW-PASS FILTER TIMECONSTANT, NSEC";TAU
198 TAUETAUE9.999999E-18
268 -
218 FOR 12=1 TO 17
228 SUMR=8
230 SUMI=0
248 FOR JX=1 TO NHARMY
250 F=PTS(1%) + J%#FREF
268 GOSUB 688
278 SUMR=SUMR + FR
288 SUMI=SUMI + FI
298 F=PTS(1%) - J/XFREF
300 GOSUB 480

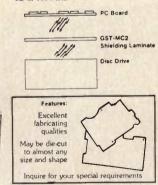
318	SUMR=SUMR + FR
328	SUMI=SUMI + FI
336	NEXT JY.
348	
358	GOSUB 688
368	
378	
388	and a state of the
488	
418	
	NDT IX
	PRINT CHRs (26)
	LPRINT "Number of harmonic terms included in summation ";NHARMC
458	LPRINT
468	LPRINT "Loop reference frequency ";FREF
	LPRINT
488	LPRINT "Type II loop Hn ":WN
498	LPRINT
588	LPRINT "Loop damping factor, Eta ";ETA
	LPRINT
528	LPRINT "Loop LPF time constant, nsec ";TAUEIE+89
	LPRINT
	LPRINT "APPROXIMATION TO SAMPLED OPEN-LOOP GAIN FUNCTION"
	LPRINT • Continuous Gol*
	LPRINT " Summation of Gol Terms Exp( +ST/2 )"
	LPRINT
	LPRINT *F Gol,dB Ang,Deg *
	LPRINT "management and and an and an and an and an and an and and
	FOR 12=1 TC ."
610	
638	
648	
650	LPRINT USING "ANARAMA ANARAMA ANARAMA ANARAMANANA ANARAMANANA
	PTS(IX) ,MAG(IX) ,ANG(IX) ,NORGAIN ,NORANGE 188/PI
	ND(T 1/2
678	STOP
	· MAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAAA
	OPEN LOOP GAIN FUNCTION, CONTINUOUS
	*#EXEXEXEXEXEXE
	W=F#2#P]
	A= (WN/W) *2 # SOR( (1+ (2#ETA#W/WN) *2) / (1 + (W#TAU) *2 ) )
	BPI + ATN(28ETA8H/M) - ATN(WETAU) FR-AECOS(B)
	FI=A#SIN(B)
	RETURN A

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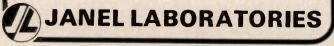
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PD7852	12	2-512	25	1.5
PD7905	4	2-50	30	1.2
PD7848	8	800-960	25	1.35

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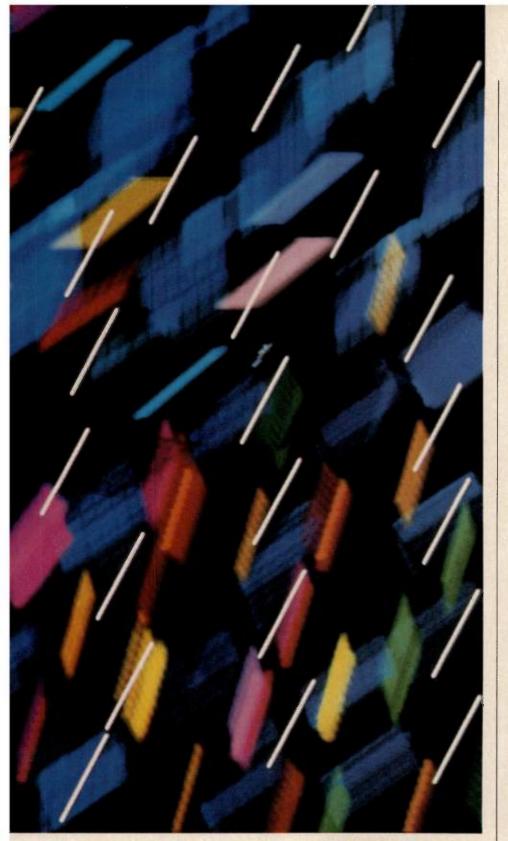
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We began with Magnetic Field Instruments and Radio Frequency Antennas for the military. Since the mid-sixties we have been working closely with the military and industry standard setting groups, and EMCO now has one of the broadest lines of Test Antennas and EMI/RFI Test Accessories in the World.

### What exactly is your major product line?

EMCO's primary business is Test Antennas for use in Emissions and Immunity (Susceptibility) Testing as required for MIL Standard, FCC, VDE, and CISPR Test Procedures.

Typical Military Antennas for Radiated Immunity (RS) and Emissions (RE) Testing cover the frequency ranges from 30 Hz to 18 GHz, and are noted in MIL STD 462, Notice 3, Table 1. EMCO currently manufactures Magnetic Field Loops, the 41" Rod Antenna, Parallel Strip Line, both



Biconical Antennas, the Conical Log Spirals and the Double Ridged Guide Antennas shown on this table.

Antennas which are currently acceptable for use in FCC Volume II, Part 15 Emissions Testing include, Adjustable Element Dipole Sets, Broadband Biconical Antennas and Broadband Log Periodic Antennas. EMCO manufactures all of these separately or can include them as part of an FCC "Class A" and "Class B" Antenna Test System.

### What differentiates your antennas from your competitors?

One major difference is Calibration. Each Antenna is calibrated using NBS Traceable Testing Equipment, on our own FCC open field test site. Calibration data includes Antenna Factor, Numeric Power Gain, and dBi Gain for each individual Antenna. For Immunity Testing Antennas we include Field Strength measurements in Volts Per Meter, and Radiation Patterns where applicable.

Another difference is Design and Construction. Each Antenna is designed to be durable and long-lasting, yet functional in varied applications, such as in Anechoic Chambers or Outside Test Sites. Antennas and accessories are machined and constructed "in-house" for Optimum Quality Control.



One last difference and maybe the most important, is EMCO's continual Product Personarch and De-

improvement thru Research and Development. For example, our **Dipole** and **Biconical Balun** design is much improved from the old DM-105 and military designs . . . and we are continually researching and redesigning to make EMI/RFI Testing simpler and more accurate.

### What other Test Equipment and Accessories do you offer?

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# "Hush-Hush" Technology Comes Out of the Closet

### Keene Corp., Shielding Division

**Generative Science Construction Security of government facilities, is coming out of the closet. Hastening its "coming out" is the boom in consumer electronics, medicine's fast-growing MRI\* (Magnetic Resonance Imaging) miracle machine, and an expected change of heart by the FCC.** 

"The market has grown 25% a year since '79 and the pace is increasing," says Arnold Zais of Keene Corporation's Shielding Division. "In '84, the total market should top the \$200 million mark. Our own sales should be up 20% this year."

Sometimes built invisibly within walls so outsiders don't know it's there, RF shielding blocks passage of radio waves like insulation blocks heat loss in your home. So it stifles deliberate electronic bugging and acidental radio interference. In RF shielded rooms, even the power lines and air vents are specially treated so stray radio waves don't pass through.

"For years, our main customers were government agencies and the aerospace industry, for security reasons," says Zais. "And secondarily, from industry for quality control testing."

But in the past five years, this has completely changed around. "Non-government/military sales now represent the majority of the business and is the fastestgrowing segment," says Zais. "Part of the reason is that 'electronic hash' is growing at 50% a year.

"Literally, there's no end in sight unless people like you and me give up on computers, video games and our electronic way of life."

One main impetus for growth of RF shielding is MRI. That's the \$1 million-acopy medical "imaging" tool that's said to be the most powerful diagnostic device ever developed. Some observers expect it to eclipse X-rays and CAT scanners. It shows more and eliminates the radiation hazard.

Every MRI must be installed in a shielded room to get high quality images. Recently, for instance, the University of Pennsylvania started up the nation's most advanced MRI clinical research facility all enveloped in a \$1.5 million special room cocooned with \$100,000 worth of Keene shielding.

"Our MRI business tripled in '83," says

Zais. "And this year, the FDA is expected to approve MRI as an accepted imaging modality. As a result, this business segment, which has nothing to do with security or defense, should continue to grow exceptionally for the forseeable future."

# New FCC Product Safety Test

Also contributing to Keene Shielding's growth is the continuing boom in electronics and computers. Reason: before any electronic device can be marketed in the U.S., the FCC requires a test to ensure that it doesn't give off harmful or interfering emissions. Europe's Electromagnetic Compatibility Compliance testing requirements parallel the FCC's.

"Electronic manufacturers run emissions tests for their own purposes as well," says Zais. "They want to make sure their new product won't be 'zapped by a passerby with a CB or an automatic garage door opener or a local TV transmitter or a faulty arcing neon sign."

This year, many industry observers expect the FCC to change its longstanding regulation and give shielded anechoic test chambers a shot in the arm. For years, the FCC mandated "open site" testing for compliance testing — tests run out in the country, away from everything, where the environment is electronically "clear" but generally not. Recently, the major computer and electronics firms are buying shielded anechoic chambers for FCC testing. Although such chambers from Keene cost over \$1 million they are more efficient, reliable and accurate than outdoor sites.

### **Everybody Wins**

"This would benefit everybody," says Zais. "Shielded chamber testing is obviously more efficient and economical than trekking out to the desert with a truckload of equipment. Ironically, it also creates an electronically 'cleaner' environment because there's so much electronic clutter everywhere outdoors. So chamber test results are more reliable. I believe the FCC is simply letting the evidence and experience accumulate. And now that it has, they'll probably accept chamber testing as fully equivalent to open site testing. It



A typical testing application in a Keene shielded chamber.

could be much more cost effective for the electronics industry, with no tradeoff whatsoever to anybody. It's simply an idea whose time has come."

### Invisible Intruder, Invisible Barrier

Electronic RF shielding consists of constructing a complete metal room of thin metal panels. The panels are walls, floor and ceiling before the decorative surface is applied. Around all doors is an arrangement of interlocking metal linings that look like brass weatherstripping. Special RF filters go into all electrical and power lines. Air vents are designed to keep radio waves from getting through, too.

When the door closes on an RF-shielded room or chamber, the interior is cocooned in a continuous film of sheet metal. If there's even the tiniest breach, unwanted radio waves can leak through just like liquid finds the smallest leak in any container. MRI shielded rooms are made of special non-magnetic materials to block the radio waves with a structure that will not affect the MRI magnet field.

"Concern over electronic pollution, and even industrial espionage, has prompted some executives to cocoon their R&D labs, conference rooms, and computer centers in RF shielding," says Zais. One hotel in the Washington, DC area even considered shielding some meeting rooms to offer "something extra" in a security conscious age.

Keene's Shielding Division is also benefitting from development of "stealth" technology. Their recently-acquired subsidiary, Advanced Absorber Products, Amesbury, MA, makes absorber material that traps and attenuates radio waves.

"As long as electronics is here to stay, our 'invisible' business should do well." says Zais.

<sup>\*</sup>Editor's note: MRI (Magnetic Resonance Imaging) is the new industry terminology for what was called NMR (Nuclear Magnetic Resonance).

### rf digital connection

# Serial RS-232 to GPIB Interface

#### Erbtec Engineering, Inc.

This application note describes a method for interfacing serial RS-232 compatible signals with the IEEE-488 bus (GPIB).

This capability can be used to satisfy the needs of a variety of applications such as:

- Interface of RS-232C compatible devices
- Bus extension
- Factory data collection
- Data communications and distributed systems
- Use in noisy environments

Two approaches to baud rate selection are described. One approach provides convenient selection of any commonly used baud rate from 50 to 9600. The second method provides a single baud rate which can be set by the user to any desired value.

The user can also select number of bits, odd/even or no parity and number of stop bits via a DIP switch.

Receipt of a general or device RESET command will cause the interface circuitry to be reset.

### **Hardware Description**

The serial RS-232 to GPIB interface is implemented with a UART coupled with Erbtec's general purpose IEEE-488 bus interface module (EPI-120). The UART used in this design is the General Instruments AY-3-1015D. Any similar UART could be used in its place.

For optimum data transfer Port A of the EPI-120 is used for data input to the bus. Port B is used for parallel data transfer from the bus to the RS-232 serial output. Input handshakes are accommodated by directly tying the HSA-IN line on the EPI-120 to the Data Available line on the UART. The HSA-OUT line is tied to the RDAV line on the UART. Similarly, output handshakes are accommodated by directly tying HSB-IN to EOC and HSB-OUT to DTSTB. The RESET line on the EPI-120 is tied to the external reset (XR) on the UART through an inverter. Also the RE-MOTE line is tied to the RDE line on the UART through an inverter. Thus, receipt of a "GO TO LOCAL" command or loss of the REMOTE ENABLE signal will disable the UART.

Serial output from the UART is passed through a 1488 to establish proper RS-232 drive levels. Output from the 1488 is normally tied to pin 3 of a DB25 connector. Serial input from pin 2 is passed through a 1489 RS-232 receiver to the UART. The Request to Send (RTS) and Data Terminal Ready (DTR) lines are pulled up to +15 V through a 3.3K resistor.

The UART requires a clock pulse of 16 times the selected baud rate.

For optimum versatility a dual baud rate generator chip such as the AY-5-8116 can be used. An external DIP switch permits selection of any of the commonly used baud rates. Separate rates can be selected for receive and for transmit.

The second method uses a 555 timer chip as a simple clock generator. It generates a single common clock for both receive and transmit. Clock frequency must be adjusted by the user to 16 times the desired baud rate.

### Software Description

A sample program which illustrates how very simple it can be to communicate between a bus control computer and an RS-232 terminal device is provided in Figure 1. The sample program is written using HP's BASIC for the HP85.

Line 130 sends the string A\$ to device 01 on the Bus (select code 7). Assuming the EPI has its address select switches set to 1 it will be configured by the command string A\$ as follows:

A1 — Accept data from Port A in 8 bit format and transfer to the GPIB unchanged A9 — Interpret Port A logic levels as high true

B2 — Transfer data from the GPIB to Port B in 8 bit binary format.

- B9 Interpret Port B logic levels as high true
- C4 HSA-IN is high true; HSA-OUT is low true
- D4 HSB-IN is high true; HSA-OUT is low true
- N2 End of transfer condition signalled by two characters (when not specified they will default to carriage return and line feed)

Y1 — Enable next sequential address feature (Data on device address +1)

The command at line 210 sends the string "ENTER YOUR NAME" to the terminal device. Note the address is 702. This was enabled by the Y1 command in the initialization string A\$.

The command at line 230 addresses the EPI as a talker and reads in the response from the RS-232 device.

Line 250 cuases the message USER NAME IS XXXXXX to be displayed on the bus control computer — where XXXXXX is the data entered at the remote RS-232.

### Figure 1

100 ! CONFIGURE INTERFACE 110 DIM A\$[23] 120 A\$="A1, A9, B2, B9, C4, D4, N2, Y1" 130 OUTPUT 701 USING "K"; A\$ 200 ! 210 OUTPUT 702; "ENTER YOUR NAME" 220 ! 230 ENTER 701; N\$ 240 ! 250 DISP "USER NAME IS"; N\$ 900 ! 999 END

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rf designer's notebook

# A One Transistor FM Transmitter

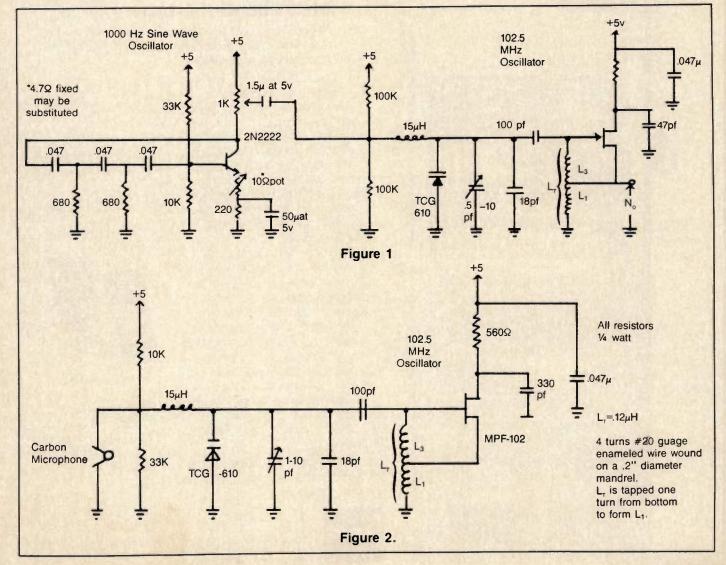
### By William Rynone Pegasus Data Systems

Last summer, I was requested to design a simple FM transmitter to meet the requirements for a convenient, low cost method of testing a large number of vendor supplied FM receivers that were to be incorporated into a digital data transmission system. The test that had to be performed was intended to make a subjective check of the accuracy of the tuning dial setting and the sensitivity of each receiver unit. Since it might become necessary to perform field tests, interconnecting various pieces of bulky and heavy test equipment was considered to be an undesirable alternative. To meet the requirement, the transistor circuit shown below was designed, built and tested. The 2N2222 circuitry is a three element phase shift oscillator circuit, designed to yield a 1000 Hz sine wave. The 1000 Hz sine wave is then applied to the TCG-610 varactor diode (6 pF at 4 volts) which changes the tank capacitance, thus varying the RF oscillator frequency at a 1000 Hz rate. The 1000 ohm potentiometer in the collector circuit may be adjusted to enable desired frequency modulation level.

The Hartley RF oscillator, which is designed around a readily available MPF-

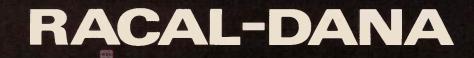
102 JFET, has an output that should be relatively stable if it is not enclosed (but without an antenna). When enclosed, a BNC or F connector can be used to feed the RF to a small loop. If you decide to build this unit, be careful. The FCC has regulations regarding the radiation of RF.

One of my colleagues, Bohden Stryzak, modified the transmitter by eliminating the sine wave oscillator portion and replaced it with a carbon microphone as shown below. His children then had a three dollar portable transmitter that could be used with any portable FM receiver as a walkie-talkie.



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Frequency Range	1.0 to 3.0 MHz	
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Quality Factor	150,000 to 300,000*	
Effective Series Resistance	35 $\Omega$ to $100\Omega^*$	
Frequency Temperature Stability ( – 30°C to + 70°C)	$\pm$ 5ppm to $\pm$ 100ppm	
Drive Level (Max)	3 to 20μW*	
Aging (First Year + 25°C)	1.0ppm (typical)	
*Depending on Frequency		

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INFO/CARD 53

Temperature (C°)

### rf product of the month

### **Power Generator**

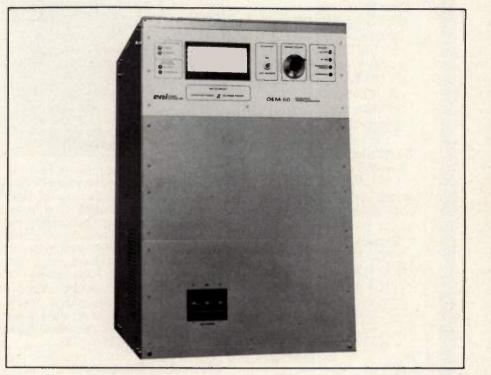
With its 5000 watt power output at 13.56 MHz, the OEM-50 is the highest power all solid state generator in the world. Although the OEM-50 is an order of magnitude smaller (5.5 cubic feet) than its vacuum tube counterparts. nothing has been left out in terms of operating features and versatility. All solid state, high reliability, accurate RF power control, fast pulse operation, stable RF power with line voltage variations, linear analog power meter with compatible digital readout, full remote control functions and built-in diagnostic test equipment are just some of the outstanding attributes of the OEM-50. If your semiconductor processing, sputtering or industrial application calls for the state-of-the-art in a 5 kilowatt source, the OEM-50 was designed just for you.

### **Automatic Power Control**

The reliable operation of any solid state power generator is directly influenced by the sophistication of its power control circuitry. The OEM-50 automatic power control module measures forward RF power, reflected RF power and the current draw of each RF power amplifier module. Should any of these parameters exceed a preset limit, the automatic power control will immediately fold back its RF output power so that the components always remain within their safe operating limits. Besides assuring safe operation of the OEM-50, the automatic power control module will provide constant RF power output level to within 3% of the matched power setting regardless of the plasma load VSWR. In addition, the automatic power control eliminates power output drift due to line voltage variations, component aging and reduces output hum and ripple to insiginficant levels. An external DC voltage or pulse fed into the rear panel connector will permit the power output of the OEM-50 to be accurately controlled by a computer program that includes end point detection information.

### **Computer Compatibility**

The OEM-50 is provided with an external computer interface bus that is compatible with TTL logic levels. This interface bus permits RF power to be turned on or off, indicates to the computer when



the unit is developing its maximum power and indicates lack of water cooling or RF power. In addition, external analog voltages are available at the interface connector for both forward and reverse power indications. These voltages are calibrated precisely at 1.00 volt per kilowatt and therefore a digital panel voltmeter will read power directly in kilowatts.

### Safe Easy Maintenance

The use of conservatively rated solid state components and automatic power control insures the user of reliable and continous performance with an absolute minimum of maintenance. A built-in diagnostic servicing switch permits the service technician to read out both the voltage and current draw of each individual module directly on the front panel meter. Should service be required, all of the plug-in modules are easily removed for replacement or repair. The very low DC voltages used in the OEM-50 greatly reduce the potential hazards associated with its servicing when compared with vacuum tube equipment.

### **Use It Anywhere**

A wide range of AC line voltages is readily accommodated by the multitap AC line transformer and the connections verified by a built-in test meter indicator. The OEM-50 may be rack mounted, using the rack mounting kit, into any 19 inch relay rack or operated remotely within the plasma system cabinetry.

### No RFI Problem

The OEM-50 is provided with an extremely well shielded and filtered power supply virtually eliminating conducted line leakage. An extremely sharp low pass filter at the output of the unit insures that all harmonics are reduced to very low levels. Extensive use of shielding and RF suppression techniques permits the unit to more than meet FCC requirements for ISM equipment at the same time that it eliminates any RF susceptibility problems for associated plasma system circuitry. ENI Power Systems, Inc., Rochester, NY 14623-2881. INFO/CARD #113.

### PROGRAMMABLE ATTENUATORS FOR OEM

### DC to 2 GHz

2, 5 or 8 cells

### Wide Selection: 0.5 dB/0.5 dB steps 1.2 dB/0.1 dB steps 12.0 dB/1.0 dB steps 31 dB/1.0 dB steps 63.75 dB/0.25 dB steps 120 dB/10 dB steps 127 dB/1 dB steps 150 dB/10 dB steps

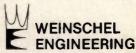
### Switching time:

6 msec @ + 12 vdc

### Rated switch life: 10<sup>7</sup> operations per cell

#### Compact size:

 $1.2 \times 1 \times 7/8 - 1$  cell 4 × 1 × 7/8 - 8 cells 3 × 1 × 7/8 - 5 cells



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### rf products

### Electro-Static Interference Simulator

The Model 2600 will simulate Electro-Statics, Electro-Magnetics, Electro-Overstress, and Electro-Impulse, in quantifi-



able, repeatable energy levels. Features include positive or negative polarity from 0-25 kV, and an adjustable ramp up, delay, and ramp down capability. The Model 2600 also features single or multiple pulse selector, pulse interval adjustment, digital voltage meter, handheld pistol grip with built-in tripod and durable carrying case with remote controller. Pulse networks for human or inanimate simulation, E and H field attachments, Ion source attachments, and Electro-Static attachments are available. IMCS Corp., Mountain View, CA 94043. INFO/CARD #137.

### **Metallized Capacitors**

Called the MDD, this series of capacitors is specifically designed for application in blocking, coupling, filter and bypass circuits of the industrial, electronics, telecommunications, and automotive industries. This resin-coated, epoxy-dipped capacitor offers cost savings over pre-



viously available molded plastic-type polyester capacitors. This new metallized polyester film features high insulation resistance, dielectric strength, and resistivity against humidity. Capacitance ranges from .01 mF to 2.2 mF are available in voltages from 100 to 630 volts DC. These are radial lead capacitors with tinplated copper wires. Copper-coated iron leads can be made available for automotive requirements. Temperature range -40°C to +85°C. Most products supplied standard with ±10% tolerance. ITT Components, Santa Ana, CA 92705. Please circle INFO/CARD #136.

### **High Speed SPDT Switch**

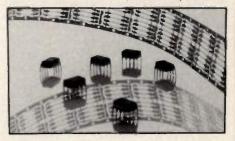
A high speed SPDT switch covers the 10-2000 MHz frequency range in a single unit. The SW-118 provides 50 dB isolation with only 0.8 dB of insertion loss, typical. The switch contains an integral TTL driver and is packaged in an 0.775" x 0.755" x



0.230" plug in package. Adams-Russell Co., Inc., Anzac Div., Burlington, MA 01803. INFO/CARD #135.

### **Optically-Coupled FET Driver**

Without requiring any additional circuitry, the new FDA-200 meets the requirements for isolated FET operations that are demanded by telecommunications, process control, data acquisition,



and ATE. In its normal work load, the FDA-200 drives either one large FET or two independent FETs that can work simultaneously, in series, or in anti-parallel for AC loads. Theta-J Corp., Wakefield, MA 01880. INFO/CARD #133.

### Miniature Quartz Crystal Resonator

Miniature quartz crystal tuning fork resonators operating in the 10-600 MHz frequency range feature low aging, high stability, low power consumption and high shock resistance. Model CX-LV resonators, designed for use in Pierce oscillators, are available in a rugged



ceramic package. Operating parameters include: temperature operating ranges — (industrial) — 40°C to 80°C; (military) — 55°C to 125°C; shock survival 1000-5000g, 1 ms, ½ sine; vibration survival 20g, rms 10-2000 Hz. Power consumption is 1mW. ETA Industries, Inc., New York, NY 10020. INFO/CARD #132.

### **High Pass Coaxial Cable Filters**

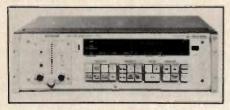
Lower frequency capabilities in semirigid coaxial cable are now possible with new High-Pass Filters. The new High-Pass Filters cut off transmissions below



Fc and allow high frequency transmissions to pass. The filters are designed for use with .141 O.D. cable. Uniform Tubes, Inc., Collegeville, PA 19426. Please circle INFO/CARD #134.

### 10 Hz to 2 GHz Vector Analyzer

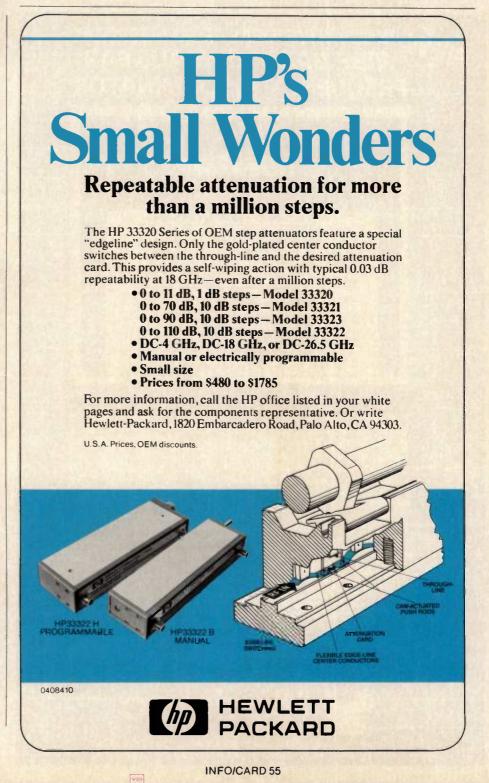
Model ZPV Vector Analyzer with 3 new modular plug-in tuners providing frequency coverage from 10 Hz to 2 GHz. Previous models provided coverage from .1 MHz to 2 GHz. The basic unit consists of



a dual-channel vector voltmeter measuring according to magnitude and phase and a microprocessor-controlled analyzer section, weighting, normalizing and converting the measured voltage vectors into the desired complex quantity. Polarad Electronics, Inc., Lake Success, NY 11042. INFO/CARD #131.

### **Active Bias Source**

The BT500 active bias thermal tracking device provides excellent thermal tracking and requires only limited external circuitry for operation. It is the first such device to be available in the standard hermetic TO-60 package; other packages are also available. The BT500 can be screened to TRW RF Devices Division's new "TX" high reliability screening specifications for MIL-S-19500 applications. The price of the BT500 is \$15.60 each in 100 piece quantities. Availability is immediate. RF Devices Division, TRW Electronic Components Group, Lawndale, CA 90260. INFO/CARD #178.





### Push-Pull High Power Bipolar Amplifier

The TPA0102-130 device provides 130 watts of CW linear power in broadband applications. Gain is over 9 dB at 162 MHz and 28 volts. The device is usable over a 30 to 175 MHz frequency range. The price of the TPA0102-130 is \$123.45 each in 100 piece quantities. They are available

now. RF Devices Division, TRW Electronic Components Group, Lawdale, CA 90260. Please circle INFO/CARD #127.

### EM Shielding for Plastics

Known as X-COAT 210-2, the new material typically covers 400 square feet per gallon at thicknesses of 1.5 to 2.5 mils, and adheres readily to most substrates.

### I-INNOVATIVE IDEAS FROM MERRIMAC-WIDE BAND, ULTRALINEAR-FREQUENCY DISCRIMINATOR 160 MHZ DISCRIMINATOR POSITIVE SLOPE 200 150 100 MILLIVOLTS INTO 50 500 0 Ω - 50 - 100 RFIN = +7 dBm - 150 - 200 100 120 130 140 150 160 170 180 190 200 210 FREQUENCY IN MHZ **Specifications:** Center Frequency\*(fc): 160 MHz \* Instantaneous Operating up to one octave Bandwidth: 50 Ω Input Impedance: 1.5:1 Max. VSWR +7 to +10 dBm **RF Input Level:** +5% over one octave Linearity: 3mV/MHz into 500Ω load **Output Level: Output Slope:** Positive Standard: (Negative on request) External fc adjustment Frequency Offset: available on request 55 to +85°C Temperature: 0.5"x1.0"x1.5" nominal Size: Connections: Pins Weight: <1 oz \* Models Available at Center Frequencies of 30, 60 and 70 MHz THE SIGNAL PROCESSING SPECIALISTS INDUSTRIES. P.O. BOX 986, 41 FAIRFIELD PLACE, WEST CALDWELL, N.J. 07007-0986 USA 201-575-1300+TWX 710-734-4314+CABLE: MERRIMAC W CALDWELL NJ

INFO/CARD 56

X-COAT 210-2 is the most versatile of the company's line of EMI shielding products. It can be cured at elevated temperatures for maximum production rates, or allowed to dry at ambient temperatures. It will adhere to a wide variety of plastics, using specially formulated thinner systems. Electrical resistance of less than one ohm per square assures excellent shielding effectiveness; X-COAT 210-2 is an Underwriters Laboratories recognized component on a wide variety of commonly used plastic substrates. X-COAT 210-2 provides very low cost shielding, as it contains no silver or other precious metals. Typical applications include instruments and computers where RF shielding and electrostatic discharge protection are required. Electro-Kinetic Systems, Inc., Boulder, CO 80302. INFO/CARD #129.

### Zero Bias Coaxial Detector/Mixer

The new RF Detector/Mixer Model DM-51-SMA/M features a replaceable, reversible zero bias Schottky diode, 10 pF output capacity, 5mV/µV sensitivity, and SMA male to SMA female connectors. It operates over the frequency range of one to 4.2 GHz, and is usable to 10 MHz with external IF bypassing. It has a VSWR of 1.35:1 nominal and 3:1 maximum at 4.2 GHz. Size is .4 in. diameter by 1.6 in. long. Elcom Systems, Inc., Boca Raton, FL 33431. INFO/CARD #130.

### **Surface Mounted Microcoil Line**

Unique to the microcoil design (designated series MC) is a solderable edge that allows visual inspection of the reflow solder joint. The new Surface Mounted Microcoils feature a high self-resonant frequency for use at greater frequencies than currently available in surface mounted coils. Standex Electronics, Cincinnati, OH 45209. INFO/CARD #128.

### Hybrid TCXO

The Model ZT-254 holds a stability of  $\pm 1 \times 10^{-6}$  over -55 to  $+85^{\circ}$ C and is packaged in a compact 1.4 x .8 x .35" (only .39 cu. in.) resistance-welded package. Many other stability vs temperature range options are also available. The oscillator is available at any frequency between 1 and 20 MHz and is designed to drive up to 10 CMOS loads. It is also compatible with TTL for 1 or 2 standard loads. The oscillator operates from an input voltage of +8.5 to +16 VDC with a typical input current of 5 mA at 5 MHz. Units plug into standard 24 PIN DIP sockets and meet an



aging rate of better than 1 ppm/yr. Center frequency adjustment is by way of an external potentiometer. Greenray Industries, Inc., Mechanicsburg, PA 17055. INFO/CARD #179.

### **Crystal Oscillators**

The XT-1000 is designed for use in microwave systems as a low phase noise reference for phased-locked oscillators. The XT-1000 comes in a rugged aluminum housing that measures 2 in. long x 2 in. wide x  $\frac{7}{6}$  in. high (51 x 51 x 22 mm). It has a flush mounting bracket enabling



easy system integration. The XT-1000 can cover 1 MHz to 125 MHz, has +7 dBm output, SMA connector and requires +6 to +15 VDC. Model XT-1000 has a stability of  $<\pm$ 15 ppm over a temperature of 0 to +55 degrees C. Model XT-1000 has  $<\pm$ 3 ppm stability over the same temperature range. Both models exhibit phase noise of >-140 dBc at 10 KHz from carrier. Microwave Research & Mfg., Upton, MA. 01568. INFO/CARD #126.

### **Detector/Preamp Hybrid**

The SD-9007 Detector/Preamp Hybrid is offered in a hermetic T0-5 package with internal feedback resistor at under \$25 in quantities over 50. The SD-9007 has a 10 kHz bandwidth and a transfer function of 3 x 10<sup>7</sup> volts/watt with 940 nm incident radiation. Custom options include bandwidth to 2 MHz and Blue or UV enhancement. Silicon Detector Corp., Newbury Park, CA 91320. INFO/CARD #125.

### **SPDT Coaxial Switch**

These SPDT latching or failsafe switches can be mounted conveniently by holes between the ports in the RF portion of the



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### THE FIRST MICROPROCESSOR CONTROLLED DISCIPLINED FREQUENCY STANDARD.

Austron's Model 2110 is the first microprocessor controlled Disciplined Frequency Standard. It offers an economical solution to the most difficult and demanding applications.

The Model 2110 automatically locks the frequency of its internal oscillator to an external reference that has superior long-term stability. Through the use of a third-order servo oscillator control technique, the instrument corrects the frequency offset and aging of the internal oscillator. Should the reference fail, the Austron 2110 can limit the frequency offset to parts in 10<sup>12</sup> for several days.

#### A SOLUTION FOR DEMANDING APPLICATIONS

- in communications systems where spectral purity and redundancy are of paramount importance
- in metrology where stable frequencies to parts in 10<sup>12</sup> accuracy are required
- in clock systems where a stable clock with accuracies to ± 100 ns is necessary

 in frequency measurement where there is a need to quickly set oscillators to very high accuracies

#### OFFERING STATE-OF-THE-ART FEATURES

- microprocessor controlled
- high-stability internal oscillator
- third-order servo oscillator control system
- frequency measurement to parts in 10<sup>-12</sup> with 100 sec averaging times
- 1 PPS clock output (externally synchronizable)
- spectrally pure output signals
- optional IEEE-488 interface
- optional dual reference frequency input allowing a choice of external references

The Austron Model 2110 microprocessor controlled Disciplined Frequency Standard is a first and it's from Austron. The 2110 can solve your most difficult timing and frequency application problems.

## CALL OR WRITE FOR MORE DETAILED INFORMATION.

### LEADERS IN TIMING AND FREQUENCY MANAGEMENT

AUSTRON INC.

P.O. BOX 14766 AUSTIN, TEXAS 78761. (512) 251-2313. TWX 910/874-1356



switch. Both inboard and outboard mounting can be provided on the same switch for increased versatility. **Teledyne Microwave, Mountain View, CA 94043. Please circle INFO/CARD #121.** 

# Remotely Controlled Signal Source

This signal source for antenna ranges or other uses can be controlled from a hand-held transmitter from distances up to 2,000 feet. The primary remote control transmitter allows selection of frequency and ON/OFF control of the signal output while the auxilliary transmitter provides control of up to 16 housekeeping functions external to the signal source. The frequency synthesized signal source can be provided to cover frequencies from 40 MHz to 200 MHz in steps as small as 10 KHz. Power output is typically 0.5 watt. FM modulation can be provided as an option. Remcon, Oceanside, CA 92054. Please circle INFO/CARD #124.

### 915 MHz Range Cavity Amplifiers

These two amplifier cavities are designed specifically as inexpensive sources of 915 MHz energy. Characterized by simple electrical and mechanical design, the EIMAC CV-2805 and CV-2811 cavities provide dependable RF power for



scientific, industrial, communication, and medical applications. Both units feature a high-gain design allowing use of solidstate drivers. The amplifier cavities are cathode driven and are forced-air cooled. Relatively compact, the cavities measure approximately 8 inches (20.3 cm.), by 5 inches (12.7 cm.), by 5 inches (12.7 cm.) Varian EIMAC, San Carlos, CA 94070. Please circle INFO/CARD #123.

### Spectrum Analyzer With "Signal Search"

Designated the MS710A, the instrument has a frequency range to 23 GHz and a wide, 100 dB, dynamic range. It is designed for easy operation and features memory presets, GP-IB interface for direct plotting, an internal preselector and dual



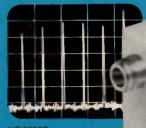
memory for trace processing. Special "signal search" functions enable rapid location of the desired signal. The PEAK-CENTER key allows the operator to zero in on the peak value and center it on the screen instantly. With other digital keys, the view can be slowly scanned left or right, or quickly shifted by half-screen increments or one entire span. Specifications include: frequency range 10 MHz to 23 GHz (in two, preselectable ranges, 10 MHz to 2 G, and from 1.7 G to 23 GHz); dynamic range 100 dB (1.7 GHz to 23 GHz), 24-pin GP-IB interface for remote operations (meets IEEE488, IEC625-1); average noise level to +30 dBm over the measuring range; resolution 1 kHz to 3 MHz in a 1, 3, 10 sequence; sweep 2 ms/div to 10 sec/div (may be selected manually or automatically coupled to frequency span, resolution bandwidth, and video bandwidth); trigger selections, single, free run, line, video and external; average noise level to +30 dBm over the measuring range; sideband noise ≤-75 dB; CRT display are 80 mm x 100 mm. Anritsu America, Inc., Oakland, NJ 07436. INFO/CARD #122.

### **Turret Attenuators**

Series 5000A and 7500A Turret Attenuators operate over a very broad frequency range and are well suited for bench setups, field use, or incorporated into test instruments. RF system applications include navigation and communication receivers. The Series 5000A Turret At-



# **Intermodulation Problems? Eliminate them** with M/A-COM



pectrum Analyzer

Solve your intermodulation problems! M/A-COM has been solving IMD problems longer than anyone in the incustry Call M/A-COM Land Mobile Communications today toll free 800-538-1533 or write



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Low Band 25-50 MHz 406-512 MHz Mid Band 66-88 MHz 806-960 MHz High Band 146-174 MHz

Low Insertion Loss

- Maximum Isolation up to 90 dB
- Single Dual, Triple Junctions
- **Constant Transmitter Impedance**
- Full Transmitter Protection 125–400 Watts

INFO/CARD 60

Designed for long lasting performance.

#### Features:

- Repairable, uncrowded circuit board with low thermal rise, long lasting performance and low production cost
- Low VSWR all ports
- Wide IF bandwidth
- Good gain stability
- Insensitive to power supply voltage variations
- Modularized for various communication bands
- Optional dual IF outputs (as shown in photo)
- Typical performance:

RF & LO Frequency (GHz) 0.6–1.1 1.0–1.7 1.7–2.7
4.4-5.0
60-80 MHz 20 dB 7 dB 6 mw 1.5/1 + 28 v @ 60 ma (+20 v optional)

Sage also offers a 70 MHz upconverter in the same package with 0 dBm output at 4.4–5.0 GHz. 70 MHz units can also be used at 40, 50 and 60 MHz IF's. Call or write for our complete product catalog and price list.



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INFO/CARD 61

**RF** Design

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State-of-the-art filter technology at its best. Only .325" H x .25" W and as short as 1.25"\*, FBC Cavity filters are setting the standard In communications, aerospace, EW, surveillance and ECM applications.

- 6 GHz-17 GHz
- 3 to 10 Poles
- Typical Butterworth amplitude, phase and group delay responses 1 dB BW of 1 to 70% of  $F_0$
- 1.5:1 or less VSWR
- As low as < 0.5 dB Insertion loss
- Up to 10 watts CW
- 50  $\Omega$  impedance
- SMA female/male connectors
- \*Length depends on number of poles and frequency

FBC cavity filters designed to your specifications within 6 to 8 weeks. Small quantities as soon as 2 weeks.



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rf products Continued

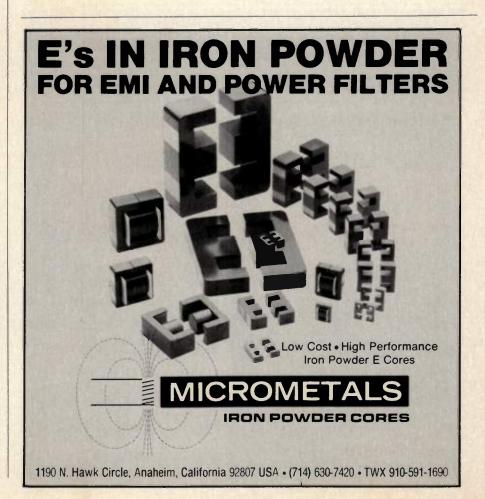
tenuators cover the DC to 2000 MHz frequency range. The positive detent action provides long life for 50 applications such as general purpose test and measurement instrumentation (both military and commercial), Two-Way Communications and general laboratory accessories. DC to 1000 MHz frequency range coverage makes the Series 7500A perfect for CATV and TV applications. Their configuration is identical to the Series 5000A; however, due to the reduction in frequency coverage, some specifications will be improved over the Series 5000A. The Series 5000A and 7500A Turret Attenuators have standard BNC female connectors. Other connectors (such as Type F or female connectors) may be ordered at no additional cost. Wavetek Indiana, Inc., RF and Microwave Components, Beech Grove, IN 46107. Please circle INFO/CARD #120.

### 1.5 GHz Sweep Generator

The LSW-359 sweep generator covers the range of 1 MHz to 1.5 GHz in three bands. The sweeper is designed for laboratory analysis of frequency versus ampli-

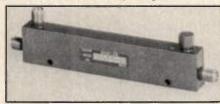


tude characteristics for communications, radar, satellite i-f, and telemetry equipment. Featuring full-band sweep, startstop sweep, delta sweep and CW operating modes, the generator also includes a versatile bypass marker system. A calibrated pulse-type marker, adjustable to any frequency within the selected band, is available in the full-band mode. In addition, birdie (comb-type) markers are available at 1, 10, 50 and 100 MHz spacings. Options include up to three spotfrequency markers and internal 1 kHz sine-wave amplitude modulation. Remote control of primary functions is easily effected, and modulation options make the unit double as a wide range signal generator. Leader Instruments Corp., Hauppauge, NY 11788. Please circle IN-FO/CARD #119.



# Broadband 10 dB Coupler

Offered is a broadband 10 dB coupler  $\pm 1$  dB (model CO518-10) with a range from 0.5 GHz to 18 GHz. Insertion loss, excluding coupling loss, is 1.2 maximum over the full frequency range. VSWR is



1.4:1 maximum, while directivity is 15 dB minimum and coupling flatness is  $\pm 0.8$ dB. The unit is 4.5" x 0.7" x .38", including SMA connectors and termination. Sage Laboratories, Inc., Natick, MA 01760. Please circle INFO/CARD #118.

# **Micro-Ohmeter**

The Model 510, a low cost, 41/2digit, micro-ohmeter is designed to measure the resistances of switch and relay contacts, transformer and motor windings, connectors, or any other low resistance devices. It has five ranges from 19.999 milliohms to 199.99 ohms, full scale, 1 micro-ohm resolution, and a basic accuracy of 0.02%. Three measurement modes are provided. The continuous DC mode is useful for making measurements on inductive components and the switched DC mode removes the effect of thermal voltages, the largest source of error in low resistance measurements. A pulsed mode is provided for thermally sensitive devices such as fuses. The standard unit comes with 4-terminal Kelvin test



clips and a parallel BCD interface. Cambridge Technology, Inc., Cambridge, MA 02140. INFO/CARD #117.

# **New Crystal Oscillator**

The XO-43 crystal oscillator features low profile, resistance welded metal pack-

Numerical Structures         Structure              Structure	Modeł Number (2)	Impedance Ohms (Power W)	Frequency Range	BNC	UNIT	PRICE (4) E	PFECTIVE	1-15-85 UHF	PC
AT-61       SOL (SW)       DC 1 SON 1       10.00       15.00       16.00				-195	123				_
AT-62         S0 (190)         DC (150)         (14.50         20.50         20.50         15.00         -         -           AT-33         S0 (1200)         DC (150)         -				14.00	20.00	20.00	16.00	-	
AT 53 AT 53 AT 55 AT 55					20.50			-	12.00
AT-54         S0 (22W)         DC-4.30Hz         -         -         -         1         5.00         -         -         -         -         1         5.00         1         - </td <td>AT-53</td> <td></td> <td></td> <td>14.00</td> <td>17.00</td> <td></td> <td>15.00</td> <td>-</td> <td>-</td>	AT-53			14.00	17.00		15.00	-	-
AT-75 or AT 800       75 or 53 (-59)       DC-1 50Hs (750MHs)       14.00       20.00       18.00       - <td></td> <td>80 (.25W)</td> <td>DC-6.2GHz</td> <td>-</td> <td>-</td> <td>-</td> <td>18.00</td> <td>-</td> <td></td>		80 (.25W)	DC-6.2GHz	-	-	-	18.00	-	
Detector, Millar, Zero Bias Scholtky:         Dif 4.20 kg         94.00         -         -         64.00         -         -           CD-31         50         .014-20 kg         94.00         -         -         64.00         -         -           Resettive Impedence Transformer, Minimum Lose Pade         -         -         -         64.00         -         -           RT-60/75         50.19.75         DC 1.50/kg         13.00         18.50         19.50         17.50         -         -           Tarminations         -         CT 53(1)         50.19.75         DC -         - </td <td></td> <td>50 (.25W)</td> <td>DC-4.2GHz</td> <td>-</td> <td>-</td> <td></td> <td>8.90 (1</td> <td>Pe2</td> <td>-</td>		50 (.25W)	DC-4.2GHz	-	-		8.90 (1	Pe2	-
CD-51         50         .014.7.0Hz         64.00         -         -         64.00         -         -         64.00         -         -         64.00         -         -         -         64.00         -	AT-75 or AT-90	75 or 93 (.5W)	DC-1.9GHz (750MHz)	14.00	20.00	20.00	18.00	-	-
Diff 31         50         014-32 million         1         -									
Resettion Impedance Transformer, Minimum Loss Pad: MT-8078         Bot 73         CC-150/Hz         10.50         19.50         19.50         17.50         -         -           Tarinston: CT-31         D </td <td></td> <td></td> <td></td> <td>54.00</td> <td>-</td> <td>-</td> <td></td> <td>-</td> <td>-</td>				54.00	-	-		-	-
R1-6073 R1-6073       B0 (8.73 B0 (8.50       C1.304z L1.50       18.50       17.50       -       -         Tarmission: CT-50 (13)       B0 (8.75 B0 (8.50)       DC.1004z L2.00       11.50       15.50       17.50       -       -         CT-51 (3)       B0 (8.90)       DC.4.204z L2.00       11.50       15.00       17.50       -       -         CT-53 (3)       B0 (8.90)       DC.4.204z L2.00       15.00       15.00       15.00       -       -       -         CT-34 (3)       B0 (8.90)       DC.4.204z L2.00       15.00       15.00       15.00       -	DM-81	90	.01-4.2GHz	-	-	-	64.00	-	-
R1-6073 R1-6073       B0 (8.73 B0 (8.50       C1.304z L1.50       18.50       17.50       -       -         Tarmission: CT-50 (13)       B0 (8.75 B0 (8.50)       DC.1004z L2.00       11.50       15.50       17.50       -       -         CT-51 (3)       B0 (8.90)       DC.4.204z L2.00       11.50       15.00       17.50       -       -         CT-53 (3)       B0 (8.90)       DC.4.204z L2.00       15.00       15.00       15.00       -       -       -         CT-34 (3)       B0 (8.90)       DC.4.204z L2.00       15.00       15.00       15.00       -	Resistive Impeda	ince Transformers, MI	Inimum Loss Pada						10
Termination:         CT 4 30W	RT-80/75	50 to 75	DC-1.5GHz	10.50	19.50	19.50	17.50	-	-
CT-50 (3)         S0 (-SW)         DC-4 20Hz         11.50         15.00         15.00         17.50         -           CT-50 (3)         S0 (-SW)         DC-4.20Hz         10.50         12.00         15.00         15.00         -         -           CT-53 (3)         S0 (-W)         DC-3.20Hz         10.50         12.00         15.00         15.00         -	M1-90/93	50 to 93	DC-1 00Hz	13.00	19.50	19.50	17.50	-	-
CT-50 (3)         S0 (-SW)         DC-4 20Hz         11.50         15.00         15.00         17.50         -           CT-50 (3)         S0 (-SW)         DC-4.20Hz         10.50         12.00         15.00         15.00         -         -           CT-53 (3)         S0 (-W)         DC-3.20Hz         10.50         12.00         15.00         15.00         -	Terminellener								
CT-61         B0 (BW)         DC-4.30Mz         9.50         12.00         12.00         18.00         15.00         -           CT-63         B0 (BW)         DC-4.30Mz         9.50         12.00         18.00         15.00         -		80 / 500	00.4.2044	44 80					100
CT-52         S0 (1W)         DC-2.50kr         10.60         15.00         16.00	CT-61	80 ( 5W)						-	-
CT-33/ML         S0 (1-W)         DC-4.20 kg         5.00 (1-W)         DC -4.20 kg         5.00 (1-W)         Image: Control of the second se	CT-52	80 (1W)	DC-2.5GHz	10.60	15.00	15.00	13.00	15.50	-
CT7.8         75 (.28/w)         DC 2.50/kiz         10.80         15.00         15.00         15.00         15.00         -           Mismatched Terrifications, 1.06.1 to 3.1; ()per Circuit, Short Circuit, Circuit, Short Circuit, Circuit, Short Circuit, Short Circuit, Circuit, Short Circuit, Circuit, Short Circuit, Circuit, Short, Circuit, Circuit, Short Circuit, Circuit, Short Circuit, Circ				5.80010	Pe) -	-	6.60(1		-
C1*93         93 (2.2947)         DC-2 SORs         13.00         18.00         1 <th1< th="">         1         <th< td=""><td>CT-54</td><td></td><td></td><td></td><td>15.00</td><td></td><td>17.50</td><td>-</td><td>-</td></th<></th1<>	CT-54				15.00		17.50	-	-
The setting         Tender         Tender <thtender< th=""> <thtender< th=""> <thtende< td=""><td>CT-93</td><td></td><td>DC-2 SGH2</td><td></td><td></td><td>15.00</td><td></td><td></td><td>-</td></thtende<></thtender<></thtender<>	CT-93		DC-2 SGH2			15.00			-
MT-31         S0         DC-100/Hz         45.50         45.50         45.50         45.50         45.50         -         -         -         -           Feed into Terminations, shunr readetor:         T         -		03 (.1.5W)	DC-2.30H2	13.00	15,00		-	15.50	-
M17.75         75         DC-1.00Hz          48.50             Press         ftm:         Terministions, shunt revelator:                   PT-00         9.3         DC-1.00Hz         10.50         19.50 </td <td>Miematched Terr</td> <td>minations, 1.06 1 to 3</td> <td>t, Open Circuit, Short Cl</td> <td></td> <td></td> <td></td> <td></td> <td></td> <td></td>	Miematched Terr	minations, 1.06 1 to 3	t, Open Circuit, Short Cl						
Deck finu Terminations, shunt relation:         Directions         Directions <thdirections< th=""> <thdirections< th="">         Directio</thdirections<></thdirections<>	MT-51			45.50	45 50	45.50	45 50	-	-
PT-50         B0         DC - L00/Hz         10.50         19.50         17.80         -         -           PT-78         P3         DC - S00/Hz         10.50         19.50         18.50         17.60         -	NT-75	75	DC-1 0GHz	-	-	45.50	-	-	-
PT-78         75         DC-SOMME         10.50         19.50 <th< td=""><td>Feed thru Termin</td><td>ations, shunt resistor</td><td>r:</td><td></td><td></td><td></td><td></td><td></td><td>1.5</td></th<>	Feed thru Termin	ations, shunt resistor	r:						1.5
PT 50         9.3         DC 1300 Hz         13.00         19.50	FT-50		DC-1.0GHz					-	-
Directional Couplex, 30 dB:         Directional Couplex, and a State Solution         Directional Couplex, and State Solution <thdirectional and="" couplex,="" sol<="" state="" td=""><td></td><td></td><td></td><td></td><td></td><td></td><td></td><td>-</td><td></td></thdirectional>								-	
DC-500         50         25.5500MHz         80.00         -			BIG-150MH2	13.00	19.50	19.80	17.50	-	-
Resisting Decoupter, series resistor of Capacities Coupler, series capacitor:         18.00         18.00         17.00         -         -           Romannia Coloma         SO         1000110000PH         DC-1.50Hz         12.00         18.00         17.00         -         -           Ch-S0 Mix 0 SMA         SO         DC-4.20Hz         -         -         13.00         13.00         -         -           Ch-S0 Mix 0 SMA         SO         DC-4.20Hz         -         -         13.00         17.00         -         -           Inductive Decoupter, series inductor:         DC-5.00Hz         12.00         18.00         18.00         17.00         -									
RD or CC-1000         1000 (1000PF)         DC-1.50Hz         12.00         18.00         17.00         -         -           CA-50 (N1 to SMA)         50         DC-1.50Hz         -         -         13.00         13.00         -         -         -           CA-50 (N1 to SMA)         50         DC-4.20Hz         -         -         13.00         13.00         -         -           Indextorts Decoupers, sets instructor:         LD-415         0.174H         DC-500MHz         12.00         18.00         17.00         -         -           Flasd Attenuator lass; 5, 10, and 20 dB, in plastic case         AT-50-857 (S)         50         DC-150Hz         -         -         AT-61 487         BO         DC-150Hz         -	DC-500	50	250-500MHz	60.00	-	-	-	-	-
Addothere:	Resistive Decoup								
CA-SO(Pito 584A)         50         DC-4/SO(Hr         -         -         10.00         13.00         -         -           Hodetty Decoupers, refs location         0.1300/HI         12.00         18.00         18.00         17.00         -         -           L0-815         0.170H         DC-500HHz         12.00         18.00         18.00         17.00         -         -           Flass Attanuator Sets. 3.6, 10, and 20 dB, in plastic cases         AT-50-857 (3.00         64.00         64.00         64.00         60.00         -         -           AT-50-857 (3.00         DC-15.00Hz         40.00         64.00         64.00         67.00         -         -           Rescitus Mutrovolares, 2 and 4 output ports:         TC-1254         50         1.5-123MHz         64.00         -         -         -         -         -           RC-320         50         DC-5.00Hz         64.00         -<	RD or CC-1000	1000 (1000PF)	DC-1.5GHz	12.00	18.00	18.00	17.00	-	-
Inductive Decembers, series inductor:									1000
LD #15         0.17/LH         DC SOOMH2         12.00         18.00         17.00         -         -           Flaed Attenuator Sets. 3.6.10, and 20 dB, in plastic case!         A         <	CA-50 (N to SMA	50	DC-4 2GHz	-	-	13.00	13.00	-	- 1
LD 4815         0.17 uH         DC 5004Hz         12.00         18.00         17.00         -         -           Flaed Attenuator Bats, 3, 6, 10, and 20 dB, in plastic case         A         A         50.00         A         0.00         40.00         40.00         40.00         -         -         -           Flaed Attenuator Bats, 3, 6, 10, and 20 dB, in plastic case         A         40.00         64.00         64.00         60.00         -         -         -           A7:05 8ET (3)         50         DC 15.00 Hz         45.00         84.00         64.00         76.00         -         -           Rescript Multicores, 2 and 4 output ponts:         TC         -         87.00         -         67.00         -         -           TC:1254         50         1.5-123MHz         64.00         -         67.00         -         -         -           Rescript Multicores, 2 and 5 output ponts:         -	Inductive Decou	niers andies inductor:							
LD 688         8.8/H         DC-558/Hz         12.00         18.00         17.00         1         -         -           Fibed Attenutor Rest, 3.6, 10, and 20 dB, inplastic cases         AT-50.687 (8)         50.00         C1.150/Hz         60.00         64.00         64.00         64.00         64.00         -         -         -           Reactive Multicorplex, 2 and 4 output ports:         TC-1252         50         1.5-125MHz         64.00         -         67.00         - </td <td>LD-R15</td> <td></td> <td></td> <td>12.00</td> <td>18.00</td> <td>18.00</td> <td>17.00</td> <td>-</td> <td>-</td>	LD-R15			12.00	18.00	18.00	17.00	-	-
AT-50-887 (9)         50         DC-1.30Hz         60.00         64.00         64.00         76.00         -           Reactive Multicouples, 2 and 4 output ports:         TC-125-2         50         1.5-123MHz         64.00         -         -         -         -           Reactive Multicouples, 2 and 4 output ports:         TC-125-2         50         1.5-123MHz         64.00         -         67.00         -	LD 6R6	Hus B						-	
AT-50-887 (9)         50         DC-1.30Hz         60.00         64.00         64.00         76.00         -           Reactive Multicouples, 2 and 4 output ports:         TC-125-2         50         1.5-123MHz         64.00         -         -         -         -           Reactive Multicouples, 2 and 4 output ports:         TC-125-2         50         1.5-123MHz         64.00         -         67.00         -	Fixed Attenuator	Sets 3 6 10 and 20	and attended in Bb (						1
AT 51 88T         B0         DC -1.50 Mz         43.00         e4.00         e4.00         e4.00         e6.00         -         -           TC 128 2         50         1.5-123 MHz         64.00         -         67.00         -         -         -           TC 128 2         50         1.5-123 MHz         64.00         -         67.00         -         -         -           Reakting Power Diviser, 3.4 and 9 ports:         -	AT-50-8ET (3)	50	OC-1.5GHz	60.00	84.00	84.00	76.00	-	-
Basettive Multicoupiers, 2 and 4 output ports:	AT-51-8ET	80	DC-1.SQHz					-	-
TC-128-2         50         1.6-123MHz         64.00         -         67.00         47.00         - <t< td=""><td>Reactive Multico</td><td>uplem, 2 and 4 output</td><td>t ports:</td><td></td><td></td><td></td><td></td><td></td><td>- X3</td></t<>	Reactive Multico	uplem, 2 and 4 output	t ports:						- X3
TC-1254         50         1.5-12344is         67.00         -         81.50         81.80         -         -           RC-320         S0         DC-2.00Hs         64.00         -         -         64.00         -         -         67.00         -         -         66.00         -	TC-125-2	50	1.5-125MHz	64.00	-	67.00	87.00	-	-
RC-3-30         80         DC-2-00.Hz         64.00         -         -         64.00         -         -         64.00         -	TC-125-4	50			-			-	-
RC-3-30         80         DC-2-00/Hz         64-00         -         -         64.00         -         -         64.00         -	Resistive Power I	Dividers, 3. 4 and 9 or	orta:						
AC:-3:0         BO         DC-6x004Hz         64.00         -         -         64.00         -         -         File         -	RC-2-30	50	DC-2.0GHz	64.00	-	-	64.00	-	-
MC:378,473         75         DC:6008Hz         64.00         -         -         64.00         -         -         64.00         -         -         54.00         -         -         54.00         -         -         54.00         -         -         54.00         -         -         54.00         -         -         54.00         -         -         54.00         -         -         -         54.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         -         34.00         -         -         <	RC-3-30		DC-BOOMHz	84.00	-	-	84.00	-	-
Ocube Islanced Misers         51.000MHz         61.00         -         71.00         61.00         -         34.00           DBM 10000         50         2:500BHz         -         -         -         -         -         34.00           DBM 5000C         50         2:500BHz         -         -         -         -         34.00           PL-50         50         0:C1.50Mz         12.00         18.00         -         17.00         -         -           PL-50         50         0:C1.50Mz         12.00         18.00         -         17.00         -         -           NOTE:         10:00 Hz         10:00 Hz         19.00         -         7.00         -         -         -         -         -         -         34.00           Bothty foodset         70         12.00         18.00         -         17.00         -			DC-SOOMHz	-	-	-		-	-
DBM: 1000         S0         S-1000MHz         S1.00         -         71.00         81.00         -         34.00           DBM: 500F         50         2:500MHz         -         -         -         34.00           AF Fues. //s Amp: and 1/16 Amp:         -         -         -         -         34.00           FF L-50         50         DC-1:50Hz         12.00         18.00         -         17.00         -           FL-53         70         DC-1:50Hz         12.00         18.00         -         17.00         -         -           FL-75         70         DC-1:50Hz         12.00         18.00         -         17.00         -         -           Schottry diodes. Mill Spec: plateid parts, and consection in nicksi, eliver, and gold. 21.56 ecisioner.         Schottry diodes Mill Spec: plateid parts, and consection in micksi, eliver, and gold. 21.56 ecisioner.         Eliver Vis diodes I           Without notice. Shipping 85.00 Domestic or 315.00 Poreign on Prepaid Orders.         Delivery is stock to 30 ays ARO.			OC-2006H1	64.00	-	-	64.00	-	-
DBM:SORC         50         2-5008Hz         -         -         -         34.00           RF Fuse, 1/8 Amp; and 1/18 Amp;         RF Fuse, 1/8 Amp; and 1/18 Amp;         -         -         -         -         -         -         34.00           RF Fuse, 1/8 Amp; and 1/18 Amp;         0         DC-1.50Hz         12.00         18.00         -         17.00         -							1.10		102
APF Fluxe. 1/5 Amp; and 1/10 Amp;         DC:1:50M1         12.00         18.00         —         17.00         —           FL-50         75         DC:1:50M1         12.00         18.00         —         17.00         —           FL-75         75         DC:1:50M1         12.00         18.00         —         17.00         —         —           FL-75         75         DC:1:50M1         12.00         18.00         —         17.00         —         —           FL-75         76         DC:1:50M14         12.00         18.00         —         17.00         —         …				61.00		71.00	61.00	-	
PL-50         50         DC-1:50H s         12.00         16.00         —         17.00         —         —           PL-75         75         DC-1:50H s         12.00         16.00         —         17.00         —         —           NOTE: 1) Critical parameters fully tested and guarniced, Fabricated from Hill, Spec, High-Rei, resistors         —         —         —         —           Monte: Specify diodes         Mill, Spec, pleted parts, and consectors in nicksi, silver, and gold, 2) See catelog for complete Model         —         … <t< td=""><td></td><td></td><td>8-000mms</td><td>-</td><td>-</td><td>-</td><td>-</td><td>-</td><td>341.00</td></t<>			8-000mms	-	-	-	-	-	341.00
FL-78 78 00-11,804 12:00 18:00 - 17:00 NOTE: 1) (Truit parameter hult) tealed and guarneed, fabricated from 411, 80xx; High-Ret, resembles tealed Schotty dicede. Mil. Spec, plated parts, and consection in inclusi, ellerer, and gold. 21.8ex catalog for complete Modal Number. Specify connector sease. Bopcular serialismi, 31 (Califordiate), and gold. 21.8ex catalog for complete Modal Number. Specify connector sease. Bopcular serialismi, 31 (Califordiate), and gold. 21.8ex catalog for complete Modal Number. Specify connector sease. Bopcular serialismi, 31 (Califordiate), and gold. 21.8ex catalog for complete Modal without notice. Shipping 85:00 Domestic or \$15.00 Foreign on Prepaid Orders. Delivery is stock to 30 days ARO.									1978
NOTE: 1) Critical parameters fully tested and guaranteed. Fabricated from Mil. Boc. High-Rei, resident Behotity diodes. Mil. Baec, plated parts, and connectors in nicksi, eliver, and gold. 2) See catalog for complete Model Number. Specify connectors areas. Bepecife versions. 31 Calibration marked on labol of unit. 4) Proc. esubject to change 1985-A without notice. Shipping 85.00 Domestic or \$18.00 Foreign on Prepaid Orders. Delivery is stock to 30 days ARO.						-	17.00	-	-
Schoffsy diodes. Mil. Spec, pletso parte, and connectors in nickel, altery, and gold. 21 See catalog for compise two doal Numbel, Specky connectors areas. Speciate available. 31 Calibration marked on lead of unit. 4) Price subject to change 1985 A arthout notice. Shipping 85.00 Domestic or \$18.00 Foreign on Prepaid Orders. Delivery is stock to 30 days ARO.						-		-	-
Numbel: Specify connector sexes: Specials evailable: 3) Calibration marked on label of unit. 4) Price subject to change 1985:4 without notice. Shipping 85:00 Domestic or \$15:00 Foreign on Prepaid Orders. Delivery is stock to 30 deys ARO.	NOTE: 1) Critical	parameters fully tests	d and guaranteed. Fabric	ated from	MIL Spec. I	High-Rel. n	relators		0
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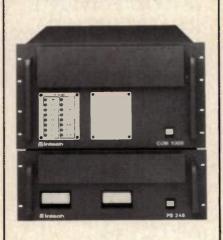
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# rf products Continued

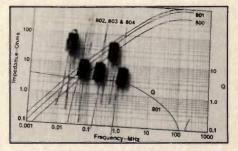
aging, with a grounding to minimize RFI. The new, hermetically sealed unit is now available at any discrete frequency be-



tween 250 kHz and 60 MHz. Models may be ordered at any of five different frequency stabilities ( $\pm$ .005%,  $\pm$ .01%,  $\pm$ .05%,  $\pm$ .1% and  $\pm$ 1%) over an operating temperature range of 0 to +70°C. Designed to withstand flow soldering without problems, the XO-43 has a hermetically sealed, all metal package (nickel plated base with stainless steel cover) which resists corrosion and provides maximum protection against humidity. The new unit is suitable for dense packaging. It has a maximum aboveboard seated height of only .225" with a length of .815" and width of .515". The XO-43 has a TTL compatible output and will drive fanout of 10 TTL loads. Input voltage requirements is +5 VDC, ±10% at 65 mZ. To minimize RFI, pin 7 is connected to ground. Dale Electronics, Inc., Tempe, AZ 85282. Please circle INFO/CARD #116.

# **Multi-Hole Ferrite Bead Chokes**

Inexpensive multi-hole ferrite bead chokes that avoid resonance problems in filtering applications have been introduced. Because low Q and other nonresonant characteristics are maintained



over a wide range with these chokes, resonance — common with air core and other choke types — is not a problem. With high resistive impedance in the



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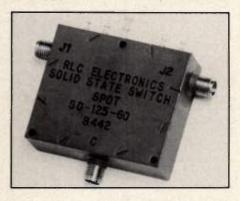
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30-300 MHz range, the chokes are effective in eliminating unwanted RF while presenting a low impedance to audio and DC. Unit costs begin at \$0.11 in production quantities. Prototypes are readily available and custom modifications can be provided. South American Development Corp., Hyde Park, NY 12538. Please circle INFO/CARD #115.

# **Diode Switches**

A new line of solid state diode switches which cover a range of .020 to 18 GHz are designed for low power and moderate speed applications. Two independent TTL



drivers are included for maximum versatility. Close tolerance printed circuit techniques and precision bonding of diodes ensure uniform operation under extreme environmental conditions. RLC Electronics, Inc., Mount Kisco, NY 10549. INFO/CARD #114.

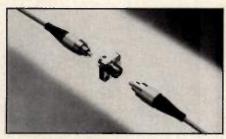
# Programmable 1 into 8 Switch

With a frequency range of DC-1000 MHz the programmable 1 into 8 switch has one common input switchable to one of eight outputs and all unused ports are internally 50 ohm terminated. Other options are avilable, for example 1 into 10 switch. The switching is accomplished by means of TO-5 type relays with a switching speed <6 msec.

This programmable switch is capable of an insertion loss of <2.0 dB at 1000 MHz, VSWR < 1.5:1 at 1000 MHz and Isolation >30 dB at 1000 MHz. It has a standard 12 volt operating voltage but is also available in 5, 6, 9, 18 & 26 volts. These are all at an operating temperature range of -55° to +71°C. Wavetek Indiana, Inc., RF and Microwave Components, Beech Grove, IN 46107. Please circle INFO/CARD #180.

# **Fiber Optic Connectors**

Single and multi-mode fiber optic connectors providing a consistent loss of less than one dB per mated pair, are now available. These connectors are FC-type Nippon Telephone and Telegraph (NTT) compatible.



They are sold without cable attached as well as in pigtails and jumpers. Connector inside diameters are available in six standard sizes ranging from 123 to 128 microns. Kyocera Industrial Ceramics Division, San Diego, CA 92123. Please circle INFO/CARD #181.

## **Toroidal Mixer Series**

This series accommodate RF and local oscillator frequencies from 1 MHz to 3 GHz. These devices are designed for 50





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ohm systems. Typical conversion loss values range from 5.0 dB to 7 dB depending upon the device. Isolation between various ports ranges typically from 30 dB to 50 dB. These devices are designed for operation from -55°C to +100°C. KDI Electronics, Whippany, NJ 07981. Please circle INFO/CARD #131.

## **Reverse Burst Acessory**

The RB-1 eliminates the long squelch tail heard with some reed type and other sub-tone decoders. When used in conjunction with decoders that offer squelch tail elimination, the RB-1 will delay



transmitter turn off time and reverse the phase of the encoded tone. This immediately stops the decoder and eliminates the squelch tail. The RB-1 is available from stock and sells for \$14.95. Communications Specialists, Inc., Orange, CA 92665-4296. INFO/CARD #130.

# **Lithium Power Cell**

The AL125 high-energy/density Lithium Thionyl Chloride Power Cell features a 1.4 amp-hr. capacity in a flat disk design that is well suited for limited-space requirements. The AL125's compact size, 1.27"x .38"/3.23 cm x .97 cm, makes it a natural choice in printed circuit design where it can be board-mounted to deliver a standard open circuit voltage of 3.6 VDC. The cell additionally features rugged stainless steel construction and a patented ceramic-to-cell seal that assures operation in hazardous environments where temperatures can range from -40°C to +70°C. With a shelf life of more than 10 years, the AL125 is unaffected by spin, altitude or position and meets the most stringent military safety standards

while offering the highest energy density at high discharge rates of any lithium cell available. Nominal capacity of the AL125 is 1.4 A-hr. at the 24-hr. rate, at 20°C, to a 2.5 V cut-off. Altus Corp., San Jose, CA 95112. INFO/CARD #159.

# EMI/RFI Shielded Instrument Case

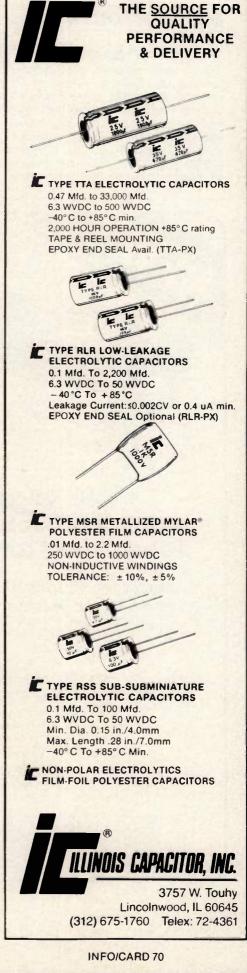
These 19-inch EMI/RFI cases are available in heights of 3 to 28 inches and depths of 16, 20 and 24 inches. Case width of 24 inches are available as modified orders. The cases achieve 55dB nominal shielding effectiveness through the use of beryllium copper gasketry. Tests performed at an independent laboratory between 30-1,000 MHz showed attenuation values at varied frequencies ranging up to 77dB. Scientific-Atlanta, Inc., Atlanta, Georgia 30348, INFO/CARD #186.

# **Optical Power Meter**

A digital optical power meter for highly accurate direct measurements, designated the ML 93A, is now available. Two optical power sensors cover a wavelength range from 0.38 to 1.8 µm and provide wide power sensitivity, from -90 to +10 dBm, with overall accuracy of ± percent. The unit features automatic zeroing, averaging function, GP-IB interface (IEEE 488), digital display, and excellent interchangability - any of seven optical power sensors can be connected without readjustment. The meter includes sensor include sensor connecting cords and is available in 120 VAC, 240 VAC models. A rechargable battery pack and charger (optional) provide up to four hours of con-



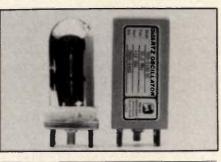
tinuous operation. A lower cost, analog unit, designated the ML 94A, and a handy, portable, digital unit, designated the ML96A are also offered. The optical power sensors for high levels (MA 97A/98A) use large active area elements for optical power reception so that the optical input can be easily modified. Available are a small active area sensor (MA911A), a slim style sensor (MA 912A) and a moderate power sensor (MA 913A) Anritsu America, Oakland, NJ 07436. Please circle INFO/CARD #185.





# Low Power Oven Oscillator

A miniature low power oven oscillator provides ultra-low phase noise for frequencies in the 3 to 11 MHz range. A vacuum flask insulator allows the "Small Fry" oscillator to consume less than 1 watt at 25°C while maintaining excellent temperature stability of  $\pm 5 \times 10^{-9}$  from -25°C to +65°C. SSB phase noise can be specified to -140 dBc (10 Hz) for 5 MHz





# 1 GHz Microprocessor-Based Intelligent Counter Offers High Resolution at Low Cost.

# Sigmotek Model ITC-3 intelligent counter offers these outstanding features:

- Wide range 0.1 Hz to 200 MHz for nonscaled input, 50 MHz to 1 GHz for prescaled input.
- High resolution over full range—for example, 10 nHz at 1 Hz, 1 Hz at 100 MHz.
- 8 digits, with at least 7 digits displayed in 1 second.
- High sensitivity—10 mV rms typical.
- Self-diagnostics.
- Precision crystal timebase or optional TCXO (  $\leq$  1 ppm, 0° C to 50° C).
- IEEE-488 (GPIB) optional.
- Compact, lightweight design.
- Low price—only \$363.00 for 1 GHz version; \$298.00 for 200 MHz version, TCXO option adds \$60.00. IEEE-488 option adds \$200.00.

Sigmotek International Corporation 4480 Enterprise St. • Fremont, CA • (415) 490-6500 oscillators and to -135 dBc for 10 MHz. 1 KHz noise is better than -165 dBc. Wenzel Associates, Inc., Austin, TX 78759. Please circle INFO/CARD #188.

# Linear Bipolar Chip

The chip, named Samson, provides a 1 A npn darlington output and a 1 A diode with 80 V breakdown voltage, two 200 mA npns, 6 general-purpose npns, three lateral pnps, one 5.8 V buried zener, one epi pinch FET, two base-emitter pinch resistors and 24 resistors (total resistance 54 kohms).

Samson replaces hybrids because it carries its own high-power output capability. It thus reduces costs, lowers the parts count and so improves reliability. Polycore Electronics Inc., Newbury Park, CA 91320. Please circle IN-FO/CARD #190.

# SSMA to SMA Adpters

The interseries adapters use the WPM-4 (Weinschel Precision Miniature) coaxial connector which incorporates



dielectric support beads at its interface to keep the internal surfaces dirt-free and to provide a rigid support for the center conductor in both male and female versions. Frequency range is from DC to 40 GHz. Maximum VSWR (per mated pair) is 1.20 to 18.0 GHz, 1.30 from 18 to 26 GHz, and 1.40 from 26 to 40 GHz. Insertion loss per mated pair is 1 dB maximum. Weinschel Engineering, Gaithersburg, MD 20877. Please circle INFO/CARD #191.

# **AC/DC Hipot Tester**

New Model HC3-AT-AD AC/DC Hipot Tester with Automatic Controls can be operated in Manual, Automatic or Automatic-Remote mode. For operator's safety all high voltage connections are on unit's rear panel allowing use of a simple



probe for lab use or a variety of optional test fixtures for production line applications. Standard input is 115 V, 50/60 Hz, 100 VA with 220 V also available. Output Voltage is 0-3000 V ac and 0-4000 V dc with a variable rate of rise of 100 to 500 V/sec, current of 5 mA and distortion of less than 5% THD. **Hipotronics, Inc.**. **Brewster, NY 10509. INFO/CARD #192**.

# **Clock Multiplier**

Models WFM 5-1400 and WFM 10-1400 high order frequency multipliers are capable of multiplying 5 or 10 MHz up to 1400 MHz or above. These multipliers are perfectly suited for NAVSTAR Clock Multipliers to multiply 10.23 MHz to 1432.2



MHz. All units are Thick Film Hybrid construction for small size and light weight, and provide a minimum output of +10 dBm with a power consumption of 150 mA at +15 to 32 VDC. Spectral Purity is excellent, with Harmonics and spurs typically >-60 dBc. Wilmanco, Northridge, CA 91324. INFO/CARD #193.

# **Digitally-Tuned UHF Preselector**

This device, part number SDTA-200/ 400-N/N, has a tunable center frequency from 200 MHz to 400 MHz with a 3 dB bandwidth of 0.4%-5.0% and rejection of 18 dB per octave. Insertion loss is dependent on the bandwidth; 2.5 dB maximum at 225 MHz with a 0.4% bandwidth and 1.0 dB maximum at 400 MHz with a 0.8%



bandwidth. Typical VSWR is 1.5:1 maximum and normal power handling capability is 50 watts. The device has a center frequency accuracy of typically .1%. The control logic interface is available in a serial format, RS-232/300 baud standard, or a parallel format using BCD, Binary, or GPIB (IEEE-488) BUS. Control voltage is 24-28 VDC, though other voltages are optional. Tuning time is typically 3-6 seconds. Standard connectors are type N. K&L Microwave Inc., Salisbury, MD 21801. INFO/CARD #194.

# 0.5 to 30 MHz Gated Amplifier

The Model 515A RF Gated Amplifier features a 500 kHz to 30 MHz frequency range and provides 3 kW RMS pulse power output from 0.5 MHz to 10 MHz, 2.5 kW from 10 to 20 MHz, and 2 kW from 20 to 30 MHz. Producing sequences of coherent, high power RF pulses with stepped or continuously variable width and



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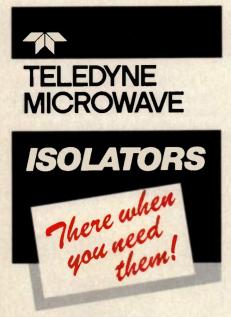
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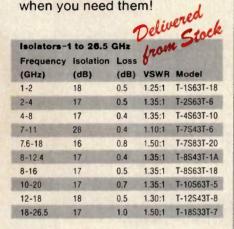
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# rf products Continued



separation, it plugs into the firm's Model 5100 Gating Modulator.

Accepting CW input levels from less than 10 mV to over 100 mV with no change in output, the Matec Model 515A RF Gated Amplifier provides an on-off ratio greater than 140 dB and a .002 maximum duty cycle for rated power output. It also accepts external pulse programs for NMR studies. Matec Instruments, Inc., Warwick, RI 02886. Please circle IN-FO/CARD #202.

## **EMI Absorber**

CHO-SERBS reduce radiated EMI for compliance with FCC and VDE emission limits, while reducing susceptibility to ESD. Unlike cable shields, CHO-SORBS do not require grounding and will not affect the data being transmitted as filters or filter pin connectors often do.

This development in EMI absorbers is made of a sleeve of specially formulated ceramic material providing a minimum of 8 dB of EMI attenuation on data and power cables in the 30-500 MHz frequency range. CHO-SORBS are available in two standard sizes, one for cables up to ¼" in diameter, and one for cables up to ¼" in diameter, and one for cable diameters up to ½". By molding a connector boot or cable jacket *over* the CHO-SORB Absorber, time and assembly costs can be saved. Chomerics Shielding Technology, Woburn, MA 01888. INFO/CARD #203.

# **Ground Plane Cable**

The new Series 65 ground plane planar cable features high density and extremely stable, electrical characteristics, making this type of cable ideal for applications in high speed computers, medical instruments, and other equipment that is sensitive to undesirable EMI radiation. Series 65 ground plane cable reduces crosstalk to 1.7% near end and 3.9% far end on a ten-foot sample with 5 ns rise time. The impedance rate is 65 ohms nominal and capacitance is 25pF/foot.

Single or double drain wire option increases the grounding capacity in the cable-to-connector interface. Cable preparation is easy and efficient. A blue tracer on the number one conductor ensures positive connector polarization. Precise center-to-center spacing on the 28 AWG cable secures the conductors to the connector with no scoring or nicking.

The Series 65 ground plane cable is available in 10 or 64 conductors and the grey PVC insulation is flame retardant, UL style 2682 recognized. Delivery is four to six weeks. Midland Ross Corp., San Jose, CA 95110. INFO/CARD #204.

## **Automatic Voltage Regulator**

New line of voltage regulators has input voltage range of  $\pm 20\%$  for  $\pm 1\%$  output accuracy. All standard service line voltages available, single and three phase, at currents ranging from 40 to 1000 amps. Optional features include individual phase control, by-pass switch, and transient suppression. **Hipotronics**, Inc., Brewster, NY 10509, please circle INFO/CARD #205.

# Ultraminiature Double Balanced Mixers

Identified as RHG "DMR" models, the compact "Hermix" series meet high density system requirement and are designed to drop in to integrated MIC assemblies. Removable connectors permit DMR models to be tested without special jigs and fixtures. Mounting centers are consistent with other industry MIC compatible mixer packages.

Model DMR8-12 is priced at \$655. Model DMR2-26 is priced at \$995. Delivery is 90 ARO. RHG Electronics Laboratory, Inc., Deer Park, N.Y. 11729. INFO/CARD #206.

# **Power Line Disturbance Meter**

The Model T1007 is a compact rugged portable instrument designed for field use. It measures the three basic parameters of an AC power line, i.e. voltage, frequency and interference. The T1007 has a built-in 4-digit interference pulse counter, 4 selectable levels of transient pulse voltage threshold (10, 50, 100 and 200 V), 50nS transient pulse capture time, and monitor of either 90-135VAC or 200-260VAC and 35-65 Hz line frequency. Tactical Electronics Corp., Melbourne, FL 32902. INFO/CARD #207.

## **Miniature RF Chokes**

Series 55X-4399 RF chokes are available in either loose piece lots or in tape mounted bandoliers for fast, automatic insertion. These miniature RF chokes offer a space saving advantage over standard molded chokes. The inductance range is from .1 microHenry to 100 microHenrys. The Series 55X-4399 RF choke is conformally coated and solvent resistant. Solder durability is 260°C in accordance with MIL-STD-810/202 and is flame retardant to UL94V-0. Color coding of these miniature devices is in accordance with MIL-STD-1285. Delivery is stock to twenty weeks. Midland-Ross Corporation, San Jose, CA 95110. IN-INFO/CARD #195.

# **Miniature Power Line Filters**

This line includes a total of 18 different series products with current ratings up to 30 amps. Included in the line are a full range of filters ranging from low cost general purpose common mode filters to premium performance multiple section units. Particularly well suited for switching power supply applications, these filters enable equipment to comply with FCC EMI/RFI regulations, along with VDE, CSA and UL safety standards.

Delivery of these products is stock to 6 weeks and they will be available shortly through SFE's franchised distributors. SFE Technologies, San Fernando Electric Division. INFO/CARD #196.

# **Semiconductor Fuse**

The 6JX fuse for 600VAC, 0.1A to 30A applications offers a low cost alternative to other semiconductor fuses of the same rating. The 6JX is manufactured to tight specifications and features plated terminals, a ceramic body and pure silver elements. The 6JX is UL listed. The 6JX features low watt loss with fast clearing times. Interrupting capacity is rated at 2,000,000 amps RMS. Carbone-Ferraz, Inc., Parsippany, N.J. 07054, please circle INFO/CARD #197.

# **Shielded Ribbon Cables**

A family of extruded shielded ribbon cables for use with SDL connectors has been developed. A precision extrusion system allows tight tolerance control on construction widths from 4 to 24 conductors. Available in 24, 26 & 28 AWG, these cables provide 100% RFI shielding as well as easy stripability for high volume applications. Phalo Corporation, Westborough, MA 01581, please circle INFO/CARD #198.

# **Miniature Oscillator**

Less than 0.8 cubic inches in volume, the dielectrically stabilized oscillator operates at a fixed frequency of 5495 ± MHz. Output power is 10 milliwatts min.; power variation with temperature is 1 dB max.; frequency stability is ±5 ppm/,C typical and ≏10 ppm/°C max. Load VSWR may be up to 2.5:1 max. (any phase) and VSWR pulling is ≏ 1.5 MHz max. Harmonic output is -30 dBc; spurious is >90 dBc; input voltage is -15 VDC and input current is 35 ma, max. Specifications are guaranteed from -54° to +85°C. Size is 1.1 x 9.5 x .68 inches (27.9 x 24.1 x 17.3 mm) nominal, excluding tuning screw, terminals and output cable. RF connectors are SMA. DC connectors are solder pins. TRAK Microwave Corporation, Tampa, FL 33614. INFO/CARD #199.

# **Isolated Power Divider**

Model IPD-65 is an economical solution to the problems of using two satellite receivers with no downconverter interaction. The IPD-65 combines a 2-way power divider with ferrite isolators to offer more than 65dB of isolation.

A feature unique to the IPD-65 is an LED indicator that allows verification of power to the LNA. Automatic LNA power switching and DC block circuits are included to simplify downconverter hookup.

Inherent reliability, quality materials, and careful design make the IPD-65 the component of choice for combining two satellite receivers. Avcom of Virginia, Inc., Richmond, VA 23236, please circle INFO/CARD #200.

# Digitally Compensated Crystal Oscillator

The DT-100 offers frequency vs. temperature stabilities approaching those of ovenized oscillators while consuming only a small fraction of the input power required to operate an oven. The DT-100 is on frequency instantly, eliminating the lengthy wait for an oven to stabilize. The DT-100 can be ordered at any frequency



between 3 MHz and 30 MHz. Over the operating temperature range of -55 to  $+85^{\circ}$ C, the frequency stability is typically  $\pm 1 \times 10^{-7}$ . ( $\pm 3 \times 10^{-7}$  max). The unit requires a power supply of +15 VDC  $\pm 10\%$  and typcially draws 13 mA of input current. (20 mA max.) Greenray Industries, Inc., Mechanicsburg, PA 17055. Please circle INFO/CARD #201.



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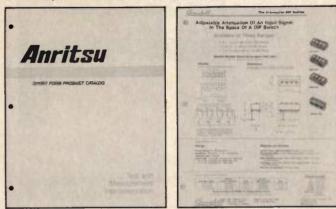
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**rf** literature

# Test & Measurement Instruments Catalog

A short-form product catalog, featuring test and measurement instrumentation for telecommunications, fiber optics, microwave and general purpose applications is now available. The 8-page catalog provides brief descriptions and general specifications on the most popular instruments in their line. Anritsu America, Oakland, NJ 07436. INFO/CARD #150.



Anritsu

Grayhill **Adjustable Attenuator DIP Switch** 

Adjustable Attenuator DIP Switches, described in new product bulletin No. 361, regulate input signals in the space of an eight position DIP switch. The switches are available in three ranges: 0.1 to 1.5 dB; 1.0 to 15 dB; and 1.5 to 22.5. Bulletin No. 361 provides complete attenuators DIP switch dimensions, electrical rating information, materials, finishes and prices. Grayhill, Inc., LaGrange, IL 60525. INFO/CARD #149.

# **High-Efficiency Power Amplifiers**

This short course brings RF design engineers up to speed on high-efficiency power-amplification techniques. Theory, practical considerations, and design procedures for amplifiers of classes A, B, C, D, E, F, and S are covered thoroughly. Envelope elimination/restoration and other techniques for combining PAs to achieve greater efficiency and/or linearity are included. Nonlinear transistor models and simulation techniques for CAD are presented. The instructor, Dr. F.H. Raab, is coauthor of Solid State Radio Engineering. The course is usually conducted at an inplant location. Green Mountain Radio Research, Winooski, VT 05404. INFO/CARD #148.

### Stop, Look, Listen

Three new booklets designed to help buyers evaluate and compare low-cost analog and digital multimeters have been published. Titled "STOP Before You Buy Another DMM," "LOOK At the Leaders" and LISTEN To Why BBC is Your Best Choice," the three booklets points out many of the factors to consider when buying a multimeter. BBC-Metrawatt/Goerz, Broomfield, CO 80020. INFO/CARD #147.

## Chip Capacitors

The Chip Capacitor Selection Guide is a 2-page, 4-color brochure describing multilayer ceramic capacitor chips. A center-



# HP 10544A/B/C or HP 10811A/B oscillators.

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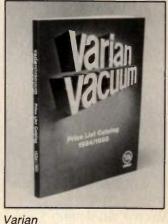
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fold chart allows simplified selection of capacitance VS chip size for 3 different families of Class I and Class II type dielectric materials. It also contains descriptions of the dielectric characteristics for each family. Johanson Dielectrics, Inc., Burbank, CA 91505. INFO/CARD #146.





Johanson

# **Vacuum Products**

The catalog includes the most current pricing information, technical information and dimension drawings for 15 vacuum product categories plus details of training curses and support services. Technical notes for nine of the product lines provide a reliable reference for applications, performance features, operation, and design. Also included is a reference section on general vacuum units and formulas. Varian, Lexington, MA 02173. Please circle INFO/CARD #145.

# **RFI/EMI Organic Coating Effectiveness**

A Product Bulletin corresponding to the chemical and physical building blocks of RFI/EMI Conductive Coatings; titled "Particle-Size Compaction." It deals with the conductive pigment types used, particle size achieved, structure and shape of particle, degree of loading and dispersion methods that vary the degree of effectiveness of an applied RFI/EMI organic shielding coatings. Advanced Coatings & Chemicals, Temple City, CA 91780. Please circle INFO/CARD #143.

# **UHF/VHF** Ground to Air Radios

This new four-color brochure describes GCA-1000 Series Ground to Air Radios for use in Military ATC Systems, Ground Control Approach systems and air Defense Tactical Communications Systems. Offering 1440 VHF channels in 116 MHz to 151.975 MHz in 25 KHz increments and 7000 useable UHF channels, 225 MHz to 399.975 MHz; coverage is enhanced by permanent memory capacity for 20 channels. Packaged in 19" rack-mount chasis the Series components, depending on mission requirements can be used together, coupled to a single antenna, as a transceiver; or seperately. Receiver sensitivity is 10 dB (s+n)/n at 3uV, 30% modulation. Transmitter output power is 25 W (VHF AM carrier) and 20 W (UHF AM carrier). Aydin Vector Division, Newtown, PA 18940. INFO/CARD #142.

# **Ultrasonics and MR Instrumentation**

A new catalog that describes a variety of instruments and accessories for performing conducting pulsed ultrasonics and magnetic resonance studies is being offered. The 6-page Short-Form Catalog of Ultrasonics and Magnetic Resonance Instruments describes a module RF gated amplifier system; broadband gated amplifiers and receivers; a modular pulse modulator



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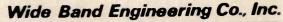


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INFO/CARD 78

# **EMI PROBLEMS?** Let Eagle Magnetic Control The Hostile Environment





rf literature Continued

and a receiver with RF pulsed oscillator plug-ins and a variety of other instruments and accessories. Matec Instruments, Inc., Warwick, RI 02886. INFO/CARD #141.





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# Vector Analyzer

This 16-page brochure describes the Model ZPV Vector Analyzer with 3 new modular plug-in tuners. Frequency coverage is 10 Hz to 2 GHz; previous models provided coverage from .1 MHz to 2 GHz. The Vector Analyzer ZPV implements a completely new, elegant technique for the measurement of complex quantities. The basic unit consists of a dual-channel vector voltmeter measuring according to magnitude and phase and microprocessor-controlled analyzer section, weighting, normalizing and converting the measured voltage vectors into the desired complex quantity. Polarad Electronics, Inc., Lake Success, NY 11042. INFO/CARD #140.

# **Coils and Inductive Components**

This colorful 28-page 81/2" x 11" catalog lists primary operational and physical parameters of a wide array of molded, shielded or unshielded coils as well as variable RF coils with inductive values from .022µH to 150,000µH. Also included is a line of toroidal inductors ranging from .10 to 10.0µH. A wide variety of Micro-i (R) miniature inductors with values from .010µH in low profile design for use on thick film Hbyrid packages are described. High current filter chokes for applications up to 15 amps and other power chokes with inductance values from 1.0 to 15,000 µH are also listed. Delevan Division, American Precision Industries, East Aurora, NY 14052-0130. INFO/CARD #139.

# Solid State Relays

A new two-page bulletin, "Solid State Relays, Series F-DC Output," is available describing the new Series F Relays which are available SPST (NO) or SPST (NC). The illustrated bulletin has schematic, mechanical arrangement, and inductive load suppression diagrams, as well as charts showing load current versus temperature for both modes. Douglas Randall, Division of Kiddee, Inc., Pawcatuck, CT 02891-0506. INFO/CARD #138.

# **Capabilities Brochure**

A 16-page full-color brochure describing the capabilities of Varian EIMAC Salt Lake is now available from the company. The division specializes in the design and manufacture of power grid tubes and switch tubes used in communications, broadcast, and defense, and its special-purpose, X-ray generating tubes for medical diagnostics applications. Varian EIMAC, Salt Lake City, UT 84104. INFO/CARD #144.

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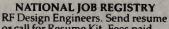
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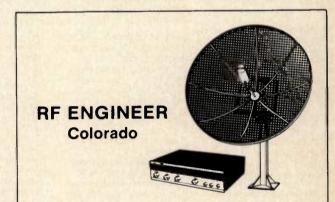
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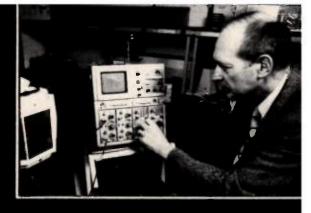
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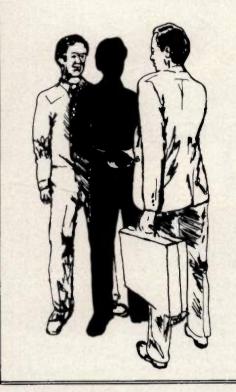
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Acrian, Inc	4
Adams-Russell - Anzac	. 33
AD-Vance Magnetics, Inc.	.75
Amperex Hicksville	.13
Amplifier Research	. 82
Andersen Laboratories,	
Inc	.15
Applied Engineering	.80
Austron, Inc.	
Avantek, Inc.	.35
Bliley Electric	.85
Carborundum	.14
CTS Knights	
California Eastern Labs, Inc.	
California Eastern Labs, Inc. Cir-Q-Tel	. 17
Cir-Q-Tel	. 17
Cir-Q-Tel	. 17 . 78 . 48
Cir-Q-Tel Cirtech Coilcraft	. 17 . 78 . 48 . 20
Cir-Q-Tel Cirtech Coilcraft Compac	17 .78 .48 .20 .81
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software	. 17 .78 .48 .20 .81 .43
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software Curtis Industries	17 .78 .48 .20 .81 .43 .11
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software Curtis Industries Daico Industries	17 .78 .48 .20 .81 .43 .11 .34
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software Curtis Industries Daico Industries Eagle Magnetic	17 78 48 20 81 43 11 34 90
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software Curtis Industries Daico Industries Eagle Magnetic EEsof	17 .78 .48 .20 .81 .43 .11 .34 .90 2
Cir-Q-Tel Cirtech Coilcraft Compac Compac Compact Software Curtis Industries Daico Industries Eagle Magnetic EEsof Elcom Systems	17 .78 .48 .20 .81 .43 .11 .34 .90 .2 .79
Cir-Q-Tel Cirtech Coilcraft Compac Compact Software Curtis Industries Daico Industries Eagle Magnetic EEsof	17 .78 .48 .20 .81 .43 .11 .34 .90 .2 .79 .60

# **Advertiser Index**

Erbtec Engineering Fischer Custom	40
Communications	55
John Fluke Mfg. Co	
Frequency Sources	
General Microwave	
Genisco	
Glasteel	57
Glasteel	73
Illinois Capacitors	.83
Instrument Specialties	
Intech, Inc	
JFW Industries	
Janel Laboratories	57
Johanson Mfg. Corp	
Kay Elemetrics	
Keene Ray Proof	.59
Log Tech	.53
M/A-Com MPD	95
M/A-Com Land Mobile	.77
Marconi	19
Matrix Systems	.10
Merrimac Industries	
Micrometals	.78
Microwave Semiconductor .	.42

Motorola Components	.28
Motorola Semiconductor	
Piezo Crystal	
Polarad Electronics	18
Programmed Test Sources	
	.89
Racal-Dana	
Repco-Telemetry	.03
SRS Industries	
Sage Laboratories	
Sawtek	
Seiko Instruments	
Sigmotek International	
Signature Antenna System	
Sokol Crystal	6
Stettner	. 56
Teledyne Microwave86,	87
Telonic Berkeley, Inc	.51
Test & Measurement World	
Texscan	
Vectron	
Voltronics	67
Weinschel Engineering	72
Wide Band Engineering	
World Business Corp.	. 50
	75
(WBC)	.75

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- Guided weapon data link amplifiers
- Military drone transmitters

## **Radar Amplifiers**

- L-band transmitters
- S-band pulse drivers for 3-D radar
- Shipboard drivers
- for AN/SPS-48 radars

### **Electronic Warfare**

- Communication jammers
- Linear AB wideband jammers
- Jamming simulators
- Expendable jammers

# MACOM

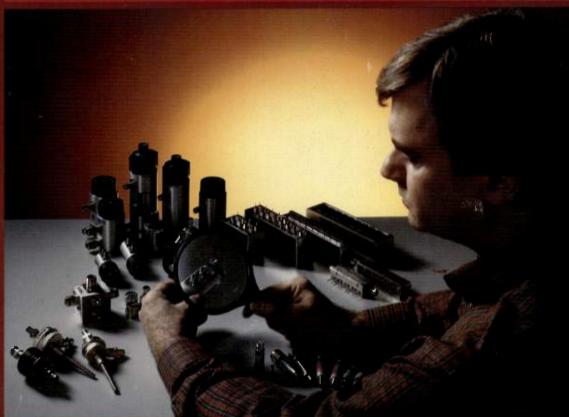
# Military Communications

- Long pulse data links
- Communication command links
- UHF transceiver amplifiers/modulators
- MILRF power boosters (ECCM)

# LaboratoryInstrumentation/Test

- Linear amplifiers
- RFI/EMI test amplifiers
- Power meter calibration systems
- Commercial and MIL power supplies
- M/A-COM MICROWAVE POWER DEVICES, INC. 330 OSER AVENUE, HAUPPAUGE, N.Y. 11788 (516) 231-1400 TWX 510-227-6239 INFO/CARD 80

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Our customers have made it a habit to buy from Texscan because they know we're not just a surface company. Quality construction, reliable performance and customer service are just a few of the underlying reasons to buy Texscan attenuators. And selection is another good reason. Choose from a wide variety of Rotary, Programmable, Fixed, Continuously Variable and Microwave attenuators, all with a choice of connectors in 50 or 75 ohm.

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