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**Special Report:** The Modular Approach to RF Design





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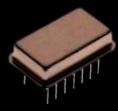
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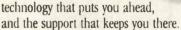
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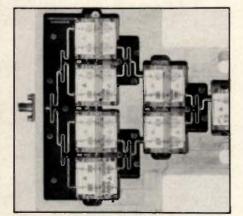
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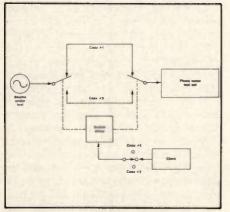
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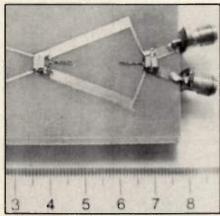
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Page 49 — Broadband Power Divider

#### **Cover Story**

#### 27 Modules Deliver the RF Punch for PLRS

Modular design is carried out to three levels in the PLRS position reporting system — amplifier assembly, three board-level subassemblies, plus thick and thin film hybrid circuits. The assembly is made for Hughes by Microwave Modules and Devices, a young company whose sole business is modular components and subsystems.

#### Features

#### 28 Special Report — RF "Building Blocks" — The New Approach to Design

Hybrids, modules and subassemblies are different names for the modular components that have become as well known to RF designers as discrete transistors and diodes. This report examines the rise of building block technology and the changes it brings to the role of the RF engineer. — Gary Breed

#### 32 Featured Technology — Specifying High Power Filters

Whether building or buying filters for high power applications, an engineer must consider the important performance factors that lie beyond passband performance. This article notes the electrical and physical characteristics that must be taken into account. — Dick Wainwright

#### 38 Phase Noise Calibrator

A runner-up prize winner in the RF Design Awards Contest, this article describes a method of calibrating and comparing phase noise measuring systems. — Guy V. Love and Kevin Lindell

#### 49 Broadband Miniature Power Splitter

This article describes a new topology for a microstrip power divider requiring minimum board space. This was another runner-up winner in the recent Design Awards Contest. — Peter Vizmuller

## 58 **RFI/EMC** Corner — Conductive Plastic Composites Offer Shielding Solution

Among the options for EMI/RFI shielding is the use of conductive plastics. This note covers some of the factors that influence a designer's choice of these materials over coatings or laminates. — Mark Gomez

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## **rf** editorial

### **'Building Block' Technology is** Coming on Strong



By Gary A. Breed Editor

With regrets, I have finally accepted the fact that the Renaissance Man is gone from the world of RF. While there are many exceptional RF engineers, the technology has advanced beyond the point where one person can master it all. As a result, the design process requires expertise from different people with knowledge of different areas — power amplifiers, oscillators, modulators, IF circuits.

However, with the present demand for RF engineers, many companies do not have all areas of design represented on their own staff. Some may choose to get by with the available talent, accepting the compromise in overall quality. Others may hire outside consultants to handle a specific design area. A rapidly growing number are buying complete subsystems from specialized companies whose purpose is to do one thing, and do it very well.

This modular approach to design and manufacturing is nothing new. The construction industry has been using preassembled doors, windows, and cabinetry for a long time. Automobiles have almost always been assembled from engines made at one plant, body panels from another, instruments from still others. "Building Block" design isn't really new to electronics, either — but it is my observation that it's the hottest thing in the RF industry right now.

#### A Shortage of Engineers? Talent Can be Purchased

There is a reason for the current high activity in RF modules, hybrids, and other supercomponents. There just aren't enough RF engineers available to meet the design needs of every company. A second reason is cost and competition. Why design a TCXO, log amplifier, or diode mixer when you can buy one that meets your specifications from several different companies that specialize in that type of design? By the time development costs are added up, it may be a lot less expensive (and certainly faster) to use a purchased module.

Of course, the drawback is *control*. Manufacturers are quite used to the purchase of individual components, but when design and manufacturing processes are purchased from suppliers instead of performed in-house, there is an understandable new concern for quality, reliability and delivery. The need for trust between the module supplier and the product manufacturer is extremely important.

Most module companies are teams of entrepreneurs with a particular type of RF expertise. They have seen the need for specialized products and are trying to use their talents both creatively and for profit. A few companies have achieved outstanding reputations, some are just starting. Like most RF companies, they are relatively small, with distinct personalities that reflect their owners and managers. Dealing with such companies is a new experience for some purchasing agents, who have only been involved with individual components from large companies.

Like every other industry, RF is evolving and changing. Because module companies are, in effect, "middle-men," they add another level to the manufacturing chain. If the present trend continues, this intermediate level could become the major force in the RF industry. We'll be watching.

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#### **rf** viewpoint

## **The Engineering Personality**



By Keith Aldrich Publisher

The subject of what kind of person becomes an engineer has been a recurring theme of sociopsychological commentary since engineering first emerged as a profession about the time of the industrial revolution.

Prior to that, the application of new scientific principles had been the sometime avocation of well-read, well-bred generalists, in much the same way that the "renaissance man" wrote sonnets and madrigals. With the "age of enlightenment," today's engineer was born, stereotypically seen as not a very social creature, more comfortable with logarithms than people, less involved with the social or economic implications of an engineering project than would seem justified to social critics.

The subject of the engineering personality and the engineer's role in project structure is again a topic of discussion, with recent announcements by MIT and other schools that they will foster a new element in the training of engineers, to develop their graduates' sense of judgment and responsibility (see Editorial, July). It is hoped that engineering expertise will thereby be brought into the highest levels of decision-making, perhaps averting such tragic and senseless decisions as the one to launch Challenger in January 1986. As a publisher who has worked with engineering audiences all his adult life without being an engineer, I feel I have some insights into the engineering personality that differ from the usual bromides. I offer them here as my contribution to the current discussion.

• In spite of their professional skepticism, engineers tend to be less jaded or cynical than most people when it comes to assessing other peoples' credibility, sincerity, and motives.

 In spite of a certain shyness about expressing it, engineers tend to feel more unspoiled, childlike enthusiasm about new experiences, places and people than many others, and about their own and others' achievements.

• The practice of engineering is much more akin to the creation of art of several sorts than is often recognized, and the engineer tends to share the artists' sensitivity to such phenomena as musical harmonics and rhythms, shades of color and light, nuances of expression, and *weights* of arguments.

• Engineers by professional training and instinct tend to be superior in sorting through spurious data to spot real facts and fairness.

These characteristics add up, it seems to me, to a relatively clear-headed kind of vision, cognizant of facts but not insensitive to sentiment, conscious of corporate or national goals but not a slave to them. And this sort of vision is, I would think, especially appropriate to the highest level of decision-making.

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SP2T	5-200	DS0492	0.5/0.8	52/40	20.0	TTL	+5, -15/30	+ 30	14 PIN DIP
SP2T	20-2000	DS0612	1.6/2.3	45/50	10.0	TTL	+ 5/3	+10	14 PIN DIP
SP2T	200-900	DS0432	0.5/0.8	45/35	1.5	TTL	+5, -15/30	+ 35	14 PIN DIP
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SP4T	10-500	DS0085	0.6/0.8	40/25	35.0	CMOS	+15/6	+10	14 PIN DIP
SP4T	500-2000	DS0259	1.3/1.5	50/25	0.4	TTL	± 5/35	+10	24 PIN DIP
SP5T	200-800	DS0475	0.6/1.0	50/40	1.0	TTL	± 5/27	+10	24 PIN DIP
SP6T	100-600	DS0506	0.6/1.0	56/50	1.0	TTL	± 5/40	+ 20	38 PIN DIP
SP8T	100-1100	DS0518	1.1/1.6	45/35	1.0	TTL	± 5/45	+13	24 PIN DIP
XFER	15-45	DS0097	.5/1.0	60/50	10.0	CMOS	+ 5/20	+10	0.8 sq 24 PIN
XFER	70-1000	DS0319	1.25/1.5	60/40	10.0	CMOS	+15/15	+10	18 PIN DIP
SPST	20-600	100C1281	1.2/1.3	80/70	0.5	TTL	+ 5/75	+10	SMA CONN
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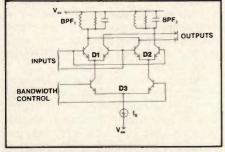
Letters should be addressed to: Editor, *RF Design*, 6300 S. Syracuse Way, Suite 650, Englewood, CO 80111.

## Variable Bandwidth IF Correction Editor:

I have discovered an error in Figure 2 of my article, "Variable Bandwidth Intermediate Frequency Amplifier," published in the June issue of *RF Design*.

I regret the typographical error and enclose the correct diagram. (Shown below — Ed.)

Steve Kuh Motorola, Inc.



Corrected version of variable bandwidth filter design.

#### Comment on Filter Design Definitions

Editor:

I thought Alan Tam's article an excellent introduction to microstrip filter design. It should be explained, however, that he elected to use a less common way of defining the cutoff frequency of his prototype. Usually the cutoff frequency of a Chebyshev filter is defined as the highest frequency where the response is  $\in$  dB down, where  $\in$  is the passband ripple. Mr. Tam's filter is designed to be 3 dB down at the cutoff frequency. This is a minor point unless someone happened to use the normalized coefficients given in the article to design another filter, which might result in an unpleasant surprise.

Methods for generating both types of coefficients (and others) are presented in Handbook of Electronic Design and Analysis Procedures Using Programmable Calculators by Bruce K. Murdock (Van Nostrand Reinhold, 1979). This book also contains much other information of interest to RF engineers.

John R. Moriarity Wavetek San Diego





#### **rf** news

#### High-Temperature Superconductors: Excitement in Materials Research

The past half year has seen dramatic changes in the theory and practical use of superconductors. As a start, the first commercial product appeared on the market (from Hypres, Inc., see April 1987 *RF Design*, p. 21) using existing lowtemperature technology. At almost the same time, news of high temperature superconducting breakthroughs began appearing in physics journals, then in the public press. Since then, every advance in research has been watched closely to see if it will be the one to bring superconductors into everyday use.

The use of superconductors in commercial applications made big news several years ago when IBM announced full-scale research into Josephson Junction (JJ) technology, with the intention of making a great leap forward in supercomputer performance. Difficulties in JJ development caused IBM to abandon its research, and superconductors were out of the news until Hypres' announcement of a picosecond-speed oscilloscope using technology based on the IBM research.

Early this year, researchers discovered new ceramic compounds that achieve superconductivity at temperatures above the 4 to 20 degrees Kelvin required for previously known superconducting materials. These materials are primarily rare-earth ceramics of the Yttrium-Barium-Copper Oxide family, which exhibit repeatable superconducting characteristics. Among the developments is the National Bureau of Standards development of a superconducting quantum interference device (SQUID) operating at 81K, above the 77K temperature of liquid nitrogen. The SQUID is the most sensitive device for measuring magnetic field, and operation using liquid nitrogen for cooling makes low cost, ultra-sensitive microwave detectors possible.

Evidence of superconductivity has been reported at temperatures as high as 280K (room temperature is 290K), although the results are not conclusive. Proof of superconductivity involves both vanishing resistivity and the Meissner effect, magnetic flux expulsion from the center of the object. In many of the recent experiments involving temperatures above 100K, the magnetic evidence (or lack thereof) suggests that the samples contain only small volumes of superconducting materials. This was reported by a group from the University of Houston, Lockheed's Palo Alto Research Lab and the National Magnet Laboratory, who observed vanishing resistivity in a ceramic oxide material at 225K. The effect survived thermal cycling over a two week period, another important test, but the small portion of the sample that was superconductive still needs to be isolated.

Superconductivity's RF applications include not only the SQUID detector, but also high efficiency magnets for oscillators and amplifiers (klystrons, magnetrons, etc.), dispersionless delay lines, low noise semiconductors, high efficiency power semiconductors and precision measurement standards.

#### **RF** Technology Assists in Plate Tectonics Measurements

Actual measurement of the movement of the earth's crust has confirmed the theory of plate tectonics, which explains how continents move in relation to one another. The June 12 issue of Science reports that radio telescopes located at Hawaii, Alaska and Japan have been used in a Very Long Baseline Interferometry (VLBI) experiment using distant guasars as "fixed" reference points. Part of the Crustal Dynamics Project, the team of radio astronomers has determined that Hawaii and Japan are moving away from each other at approximately 8.3 cm per year, and Hawaii and Alaska are being separated at a rate of 5.2 cm per year. The most remarkable aspect of the project is that results were obtained after only three years of measurements, indicating the accuracy of current VLBI techniques.

#### GE's Two Microelectronics Centers Consolidated

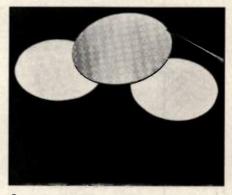
GE announced that its two microelectronic centers located in Research Triangle Park (RTP), NC, and in Somerville, NJ, are being consolidated at RTP. The move is expected to be completed by the end of the first quarter of 1988. The new GE Microelectronics Center charter will have a dual focus: to serve the application-specific integrated circuit (ASIC) requirements of GE's internal aerospace businesses, and to market its products and services with selected external military contractors and government agencies. The consolidated center will feature a complete ASIC capability, including design, CAD systems, and all the operational functions required to deliver qualified/classified parts, including customer program management and IC development.

#### Single Sideband Systems And Circuits

Single Sideband Systems and Circuits is a new book designed to bring radio communications equipment designers, electronic engineers and technicians up to date on the system and circuit design of single sideband radio communications technology. It brings together the expertise of twenty-three recognized authorities from Rockwell's Collins Radio Division to give readers the latest state-of-the-art design information, including audio and RF circuits, digital signal processing, synthesizer design, power amplifiers and many other areas. The book is edited by William E. Sabin and Edgar O. Schoenik. It is published by McGraw Hill Book Company, 11 West 19th St., New York, NY 10011. Price is listed at \$49.95.

#### **GTE Grows a Transistor**

Researchers at GTE Laboratories have developed a way to *grow* a transistor. By combining and heating silicon and the tantalum disilicide, the researchers can

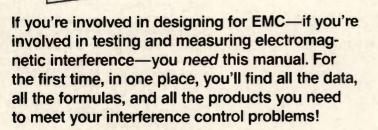


Grown transistors offer high power performance.

simultaneously grow metal connections and silicon crystals, forming a transistor. The connections in the GTE transistor penetrate the entire device, providing a three dimensional form that permits the transistor to excel in high power applications such as pulse switching.

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#### National Security Agency Endorses Rockwell

Rockwell International Corp. has received the National Security Agency's endorsement for a monolithic randomizer integrated circuit. The randomizer is the first device produced under the agency's overtake project in conjunction with its commercial COMSEC endorsement program (CCEP). The device is a digital, monolithic, binary, number generating device that can be used wherever a high speed, random data stream is required and small size and a single drive voltage are desired. It has been tested and found to have a zero/one distribution statistically unbiased making it appropriate for use in Type 1 products which secure unclassified US government information. More information can be obtained by calling Rockwell International Corp. at (714) 762-2622.

## Introducing a small accomplishment in UHF frequency sources!



Our SAW-stabilized frequency sources provide a unique solution to your demanding UHF system requirements. They pack the performance of a fine cavity oscillator into less than a tenth of a cubic inch. Their small size, low power consumption and excellent reliability are made possible by our advanced UHF Quartz SAW technology.

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temperature range and offer testing and screening to a variety of MIL Standards.

Our SAW-stabilized UHF frequency sources are being used in IFF systems, radar frequency synthesizers, GPS receivers, emergency location transmitters, fiber-optic communications and a host of other UHF and microwave system applications.

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#### Applied Computation Electromagnetics Society (ACES)

Over the past ten years, computer modeling has matured as a problem solving tool in real world electromagnetic applications. Use of computer modeling has become so routine that sessions at the various professional society meetings deal primarily with new developments. As a result, information exchange is inhibited. ACES has been formed to foster information exchange among computer modelers in electromagnetics. Emphasis is put on computer application issues rather than research and development. Further information on ACES can be obtained from Dr. Richard W. Adler, Code 62AB, Naval Postgraduate School, Monterey, CA 93943. Tel: (408) 646-2352.

## Hughes, MBB, and Aerospatiale Join to Bid on Air Defense Program

Derivatives of the battle proven Roland system have been entered by a U.S., French, and West German joint venture for the U.S. Army's Forward Area Air Defense, Line of Sight, Heavy program. Called the Paladin weapon system, the units will incorporate elements of the U.S. Roland air defense system and other components from the European Roland now fielded with the armed forces of France, West Germany and seven other nations. The Western Alliance Air Defense joint venture has been formed by Hughes Aircraft of the U. S., Messerschmitt-Boelkow-Blohm (MBB) of West Germany and SNI Aerospatiale of France to propose the Paladin. Plans call for 50 to 70 percent of the contents of Paladin to be built in the U.S. with the remainder coming from Europe.

#### **Power MMIC Services**

Microwave Semiconductor Corporation has announced the availability of GaAs custom circuit design and fabrication services. The services are tailored to meet the demands of customers requiring specialized microwave performance. The foundry services offered use the technology implemented in MSC's GaAs Power FET and standard MMIC line.

#### Electrospace and Chrysler Announce Merger

Chrysler Corporation and Electrospace Systems, Inc. announced that they have reached a merger agreement under which Chrysler will make an offer of \$27.00 per share for 100 percent of the stock of the defense contractor. The total acquisition of the stock will cost approximately \$376 million. Chrysler has been granted options by Electrospace Systems' four founding shareholders to acquire their approximately 38 percent of the company's outstanding shares. Electrospace Systems designs, develops and installs communications and electronic systems and equipment for military and commercial customers.

#### Watkins-Johnson Acquires **Microwave Center**

Watkins-Johnson Company has completed its acquisition of the Santa Barbara Microwave Center from Honeywell, Inc. W-J paid approximately \$2 million for the manufacturer of millimeter-wave mixers and oscillators.

#### **AMP and Micro-Coax Join Forces**

AMP, Inc., of Harrisburg, PA., and Micro-Coax™ Components, Inc. of Collegeville, PA., have announced agreement to manufacture and market semi-rigid cable assemblies. The assemblies will utilize AMP's semi-rigid crimp Coaxicon™ connectors and Micro-Coax Components' UT<sup>TM</sup> semi-rigid cable. Available size range of the cable assemblies include cable diameters from 0.0865 inch to 0.141 inch with frequencies ranging from DC to 26 GHz.

#### **Standard Oil Announces New** Division

Standard Oil Engineered Materials Company announced the formation of an Electronic Ceramic Division, headquartered in Niagara Falls, NY, to focus on the development, production and marketing of products. The division will invest \$8.5 million in two facilities to produce specialty materials for the microelectronic industry, specifically substrates and planar diffusion sources.

#### **Esterline Sells Product Line**

Esterline Angus Instrument Corporation, a subsidiary of Esterline Corporation of Connecticut announced the sale of its Minigraph® analog and event recorder product line to the Dickson Company of Illinois for an undisclosed sum.

#### **General Motors and Siliconix** Sian Aareement

The Delco Remy Division of General Motors, along with General Motors Research Laboratories, has signed a contract with Siliconix to develop and produce a new semiconductor power technology. According to Siliconix, the significance of this technology may be compared to MOSFET technology. It is expected to be the basis of a whole family of new devices,

both discrete and integrated circuits, promising major improvements and device efficiency at low cost.

#### Shipboard DF System Delivered to Paramax

**Electromagnetic scientists at Southwest** Research Institute (SwRI) have designed and developed an advanced shipboard tactical direction finding system that is claimed to be faster than existing systems and is strong enough to withstand the stresses of prolonged service atop a ship's mast. The first electronic support measures (ESM) system, designated the AN/SRD-502 was recently delivered under subcontract to Paramax, Inc., a Canadian defense firm, for installation aboard a new class of Canadian Patrol Frigates. It consists of a 16 foot tall antenna, a control unit, an operator unit, and data processing equipment. rf



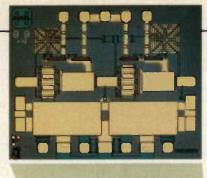


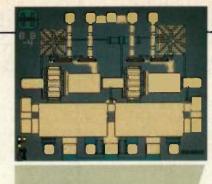
## **EMI Filters and Filter Caps**

These miniature EMI ceramic filters and capacitors are designed to suppress unwanted EMI in applications where small size is critical. Tusonix filters and filter caps cover a variety of voltage, attenuation and capacitance ranges in both solder and bushing mount styles. We have QPL approval and most are available from stock, ready for immediate shipment.

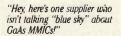
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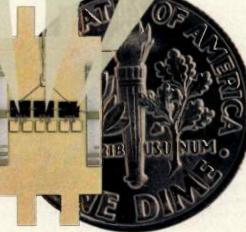










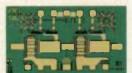


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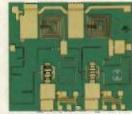
Take our little HMR-10502 GaAs MMIC amp. It provides 10



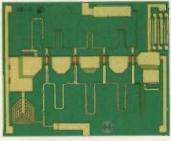
HMR-10502 MMIC



HMR-10503 MMIC



HMM-10610 MMIC



HMM-11810 MMIC MMICs shown approximately twenty times actual size. dB typical gain and 10 mW output power from 0.5 to 5 GHz.

Or its companion, our HMR-10503. It provides 10 dB typical gain and 10 mW output power from 1 to 5 GHz and includes on-chip decoupling capacitors.

If you need more power and gain, the HMM-10610 is the right choice. It offers 75 mW output power from 2 to 6 GHz with 12 dB gain.

For higher-frequency applications, we offer the HMM-11810 — with 5 dB typical gain and 50 mW output power over the 6 to 18 GHz band.

Since they're directly cascadable, you can combine multiple MMIC chips on the same carrier. They require no tuning, and minimal assembly.

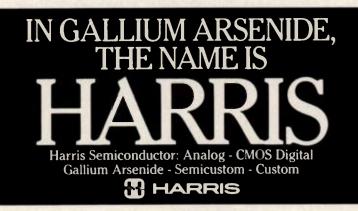
Like all Harris GaAs products, these MMICs offer improved radiation hardness over silicon devices, and are well-suited

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## **rf** calendar

#### August 25-27, 1987

#### 9th Quartz Devices Conference

Westin Crown Center, Kansas City, MO Information: Components Group, Electronic Industries Association, 2001 Eye Street, N.W., Washington, DC 20006; Tel: (202) 457-4930

#### September 1-2, 1987

Mountain States Electronics Expo Denver Merchandise Mart, Denver, CO Information: Darel A. Bennett, Midland Exposition Group, 3699 West 73rd Ave., Westminster, CO 80030; Tel: (303) 426-1634.

#### September 7-10, 1987

#### 17th European Microwave Conference

Ergife Palace Hotel, Rome, Italy Information: Microwave Exhibitors of Publishers Ltd., 90 Calverly Road, Tunbridge Wells, Kent TN12UN, England

#### September 15-17, 1987 Midcon/87

The Ohara Center, Chicago, IL Information: Midcon/87, 8110 Airport Boulevard, Los Angeles, CA 90045-3194; Tel: (213) 772-2965, (800) 421-6816, inside California (800) 262-4208

#### September 21-22, 1987

#### **1987 IEEE Bipolar Circuits and Technology Meeting** Hyatt Regency Hotel, Minneapolis, MN

Information: Janice Jopke, BCTM Secretary/Treasurer, 5016 W. 99th St., Bloomington, MN 55437; Tel: (612) 835-6742

#### September 22-24, 1987 Northcon/87

Portland Memorial Coliseum, Portland, OR Information: Show Manager, Northcon/87, 8110 Airport Boulevard, Los Angeles, CA 90045-3194; Tel: (213) 772-2965

#### September 29-October 1, 1987

**1987 OEM Design and Integrated Manufacturing Exposition** McCormick Place East, Chicago, IL Information: Bill Little, Penton Exposition, 122 East 42nd St., New York, NY 10168; Tel: (800) 634-4639, in NY call (212) 867-9191

#### September 29-October 1, 1987

Ninth Annual Meeting and Symposium, Antenna Measurement Techniques Association The Stouffer Madison Hotel, Seattle, WA Information: James R. Otley, 1987 AMTA Symposium, 6682 South 191st Place, Suite E-105, Kent, WA 98023; Tel: (206) 655-7780

#### September 29-October 1, 1987

Fall National Design Engineering Show and Conference Javits Convention Center, New York, NY Information: Monica Viladegutt, Cahners Exposition Group, 999 Summer St., P.O. Box 3833, Stamford, CT 06905; Tel: (203) 964-0000

#### October 4-8, 1987 ISA/87

Anaheim Convention Center, CA Information: Paul Albert, Instrument Society of America, P.O. Box 12277, Research Triangle Park, NC 27709

#### October 13-15, 1987

Scan-Tech 87 Bartle Hall, Kansas City, MO Information: Don Anderson, AIM, Inc., 1326 Freeport Rd., Pittsburgh, PA 15238. Tel: (412) 963-8588

#### October 13-15, 1987

**19th International SAMPE Technical Conference** Hyatt Hotel, Crystal City, VA

Information: Marge Smith, Business Director, P.O. Box 2459, Covina, CA 91722. Tel: (818) 331-0616

#### October 19-21, 1987

The Twentieth Annual Connectors and Interconnection Technology Symposium Franklin Plaza Hotel, Philadelphia, PA Information: Electronic Connector Study Group, Inc., PO. Box 167, Fort Washington, PA 19034-0167; Tel: (215) 825-3840

#### October 26-30, 1987

FOC/LAN 87, Eleventh International Fiber Optic Communication and Local Area Networks Exposition Anaheim Convention Center, Anaheim, CA Information: Renee Farrington, Information Gatekeepers, Inc.,214 Harvard Avenue, Boston, MA 02134; Tel: (617) 232-3111

#### November 10-12, 1987

**1987 ITEA Symposium** Park Plaza Hotel, Boston, MA Information: Howard L. Graves, Raytheon Company, Public Relations, 141 Spring Street, Lexington, MA 02173; Tel: (617) 470-6027





**R & B Enterprises** 

Tempest-A Detailed Design Course September 14-18, 1987, Philadelphia, PA Electromagnetic Pulse (EMP) Design and Test October 5-6, 1987, Philadelphia, PA November 17-18, 1987, Washington, DC December 7-8, 1987, Philadelphia, PA Electromagnetic Pulse Workshop October 9, 1987, Philadelphia, PA

December 11, 1987, Philadelphia, PA Understanding and Applying MIL-STD-461C October 20-21, 1987, Philadelphia, PA November 17-18, 1987, Washington, DC December 14-15, 1987, Philadelphia, PA

MIL-STD-461C Praxis (Workshop) October 20-21, 1987, Philadelphia, PA December 16-17, 1987, Philadelphia, PA

Information: Greg Gore, Director of Training, R & B Enterprises, 20 Clipper Road, West Conshohocken, PA 19428. Tel: (215) 825-1960

#### **Norand Corporation**

 FCC, VDE, CISPR, and SAE Regulation and Design Criteria September 1-2, 1987, Cedar Rapids, IA
 Grounding, Bonding, Shielding and PCB Design for EMI/EMC
 September 15-16, 1987, Cedar Rapids, IA
 MIL-STD 461B/C and 462 Regulations and Design Criteria October 13-14, 1987, Cedar Rapids, IA

**RF Susceptibility and ESD Testing** 

November 3-4, 1987, Cedar Rapids, IA

Information: Bev Reynoldson, Norand Corporation, 550 2nd St S.E., Cedar Rapids, IA 52401; Tel: (319) 846-2415

#### **Compliance Engineering**

Compliance Seminars: EMI, Safety, ESD, Telecom August 25-28, 1987, Atlanta, GA September 29-October 2, 1987, Orlando, FL October 27-30, 1987, Boston, MA December 1-4, 1987, San Diego, CA Information: Compliance Engineering, 593 Massachusetts Avenue, Boxborough, MA 01719. Tel: (617) 264-4208

#### Avantek, Inc.

RF & Microwave Semiconductor Design Solutions Seminar September 15, 1987, Santa Clara, CA September 17, 1987, Phoenix, AZ September 21, 1987, Denver, CO September 28, 1987, Kansas City, MO October 2, 1987, Ft. Wayne, IN October 5, 1987, Boston, MA October 7, 1987, Ft. Washington, PA October 9, 1987, Atlanta, GA October 13, 1987, Orlando, FL October 15, 1987, Dallas, TX October 19, 1987, Burbank, CA November 4, 1987, Albuquerque, NM

Information: Janice Little, Avantek, Inc., M/S 3G, 3175 Bowers Ave., Santa Clara, CA 95054. Tel: (408) 970-2139

The George Washington University **Frequency Hopping Signals and Systems** September 9-11, 1987, Washington, DC **HF** Communications Technology September 14-18, 1987, Orlando, FL November 16-20, 1987, Washington, DC Spread Spectrum Communication Systems September 21-25, 1987, Washington, DC Fiber Optics Technology for Communication September 22-24, 1987, Washington, DC Introduction to Modulators and Transmitters October 5-6, 1987, Washington, DC **High Frequency Spectrum:** New concepts and technologies October 5-9, 1987, Washington, DC Introduction to Receivers October 19-20, 1897, Washington, DC **Modern Receiver Design** October 21-23, 1987, Washington, DC **Global Positioning System: Principles and Practice** November 4-6, 1987, Washington, DC Wideband Communications Systems December 7-11, 1987, Washington, DC Information: Shirley Forlenzo, Continuing Education Program, George Washington University, Washington, DC 20052; Tel: (800) 424-9773, (202) 994-8530

#### Interference Control Technologies, Inc. Practical EMI Fixes

September 22-25, 1987, San Diego, CA MIL-STD-461/462

September 29-October 2, 1987, Washington, DC Information: Penny Caran, Registrar, Interference Control Technologies, Inc., State Route 625, P.O. Box D, Gainsville, VA 22056; Tel: (703) 347-0030

#### **Integrated Computer Systems**

Digital Signal Processing September 22-25, 1987, San Diego, CA October 27-30, 1987, Anaheim, CA December 1-4, 1987, Washington, DC December 8-11, 1987, Boston, MA

Machine Vision and Image Recognition September 15-18, 1987, Los Angeles, CA September 29-October 2, Washington, DC October 20-23, 1987, Palo Alto, CA December 8-11, 1987, Toronto, Canada Fiber Optic Communication September 15-18, 1987, Washington, DC September 29-October 2, 1987, San Diego, CA

November 3-6, 1987, Los Angeles, CA Hands-On Programming in C September 22-25, 1987, Washington, DC

September 29-October 2, 1987, San Diego, CA October 27-30, 1987, Boston, MA

Information: Marilyn Martin, Integrated Computer Systems, 5800 Hannum Avenue, P.O. Box 3614, Culver City, CA 90321-3614; Tel: (800) 421-8166, (213) 417-8888

## New! ¿ Symmetrical Shielding Strips (S<sup>3</sup>) provide bi-directional engagement at severe shear angles!

These new symmetrical slotted shielding strips of beryllium copper permit continuous spring contact throughout their length, providing the perfect answer for a variety of shielding requirements.

Three models are available: basic, rivet-mount and double-faced adhesive-mount designs. The basic design consists of low-compression, adhesivemounted strips. A generous radius profile provides for the greatest incident engagement angle with the lowest force. As with all Sticky Fingers<sup>®</sup> shielding strips, the self-adhesive tape makes mounting easy and secure.

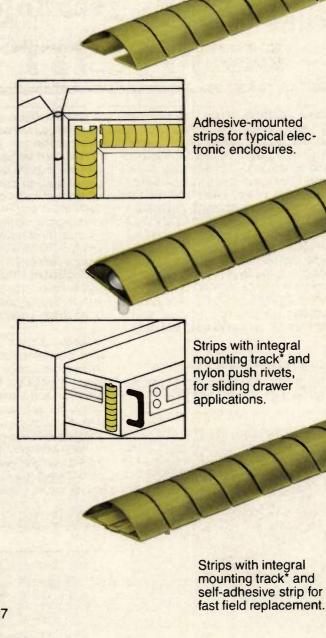
The rivet-mount design incorporates the addition of an integral track, pierced for mounting with nylon push rivets. This configuration allows bidirectional engagement, and is specially designed for slide applications, PC board connections, etc.

The third design also incorporates an integral track-mount design, but employs a double-faced adhesive tape instead of push rivets. This provides for fast, easy field replacement in military applications, especially where high frequencies do not permit the use of mounting holes.

For complete information, including exact specifications, dimensional drawings, etc., on these and other Instrument Specialties shielding strips, use this publication's Reader Service Card. Or write to us directly at Dept. RFD-36.



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TYPICAL STAN	DARD MODELS
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NC 502	up to 1 GHz
NC 503	up to 2 GHz
NC 504	up to 3 GHz
NC 505	up to 4 GHz
NC 506	up to 5 GHz

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NC 2102	up to 100 kHz
NC 2103	up to 500 kHz
NC 2104	up to 1 MHz
NC 2105	up to 10 MHz
NC 2106	up to 20 MHz
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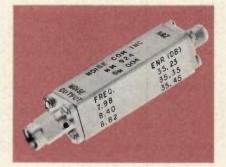


TYPICAL STAN	DARD MODELS
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NC 1107A	up to 100 MHz
NC 1108A	up to 500 MHz
NC 1109A	up to 1 GHz
NC 1110A	up to 1.5 GHz

Other frequency ranges and output levels available

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When low cost, lot-to-lot C-V repeatability and prompt delivery are critical, M/A-COM's HATVs deliver. Our HATVs provide the DEPENDABLE product performance For more information, write to: M/A-COM Semiconductor Products, Inc., South the marketplace has been waiting for.

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# H PRESENTS

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Microwave Synthesizer; Moderate-Low Noise & 1 to 10 MHz Steps
Microwave Synthesizer; Low Noise, Small Size & Low Cost-10 kHz Step
Crystal Oscillators; Free Running or Locked to 5 or 10 MHz
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Converts Magnetron RADAR to MTI RADAR >40 dB Cancellation

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INFO/CARD 19

## **Modules Deliver the RF Punch for PLRS**

Assemblies, Boards and Hybrids — Three Levels of Modularity

The Automatic Position Location Reporting System (PLRS) is a UHF radio network consisting of a transportable Master Station with computers and situation display, in communication with up to 400 Individual User Units. These units may be deployed as manpacks or installed in aircraft and surface vehicles. The Master Station computes and continuously updates the position of each User Unit within 15 meters. The PLRS system is currently supplied to the Army and Marines.

Reliable communication of position data is realized by a unique combination of an anti-jam radio and a sophisticated net management algorithm. To support hundreds of users in real-time, the radio uses a spread-spectrum, pseudonoise, frequency-hopped waveform with error detection and correction. The net management algorithm monitors all radio links and maintains real-time information on the connectivity. This data is used to reconfigure the network when necessary, as a result of a broken link.

The 100 watt Transmitter Assembly used in the PLRS radio is made by Microwave Modules and Devices (MMD) for Hughes, the system manufacturer. The assembly is an excellent example of how modules and modular construction can be applied to high volume production military hardware. The Transmitter assembly built at MMD consists of three replaceable modules, each with a designated function:

• Control Module — This module selects appropriate power output with greater than 23 dB dynamic range, controls both the rise and fall time of the pulse, and protects the amplifier module.

• Power Amplifier Module — The UHF PA delivers 100 watts, or another selected level, over the full military temperature range. A PIN diode T/R switch is included in this module.

• Driver Module — This module delivers 10 watts of power to the PA module. It is constructed as a combination of

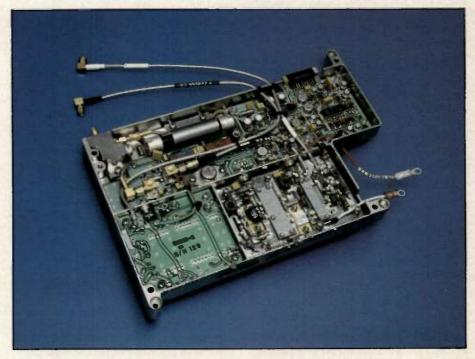
TO39 thin film hybrids and a large thick film plug-in hybrid.

The modules represent an integration of thin film, thick film and printed board technology to achieve an optimum cost/ performance tradeoff. Carrying the modular concept even farther, Hughes is able to use the assembly as a power amplifier in other radio systems operating in the military UHF band.

MMD supplies modular RF and microwave power amplifiers for defense electronics and commercial applications. They have delivered amplifiers for communications, radar, avionics and EW systems from DC to 4 GHz, operating from less than a watt to many kilowatts. Microwave Modules and Devices is located at 550 Ellis Street, Mountain View, CA 94043, tel. (415) 961-1473.



The driver subassembly combines thick and thin film technology.



The amplifier assembly. At the top is the control module, lower left is the driver, and at lower right the power amplifier.

### rf special report

## RF "Building Blocks" — The New Approach to Design

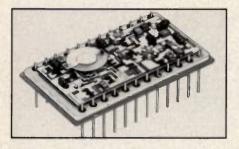
#### New Super-Components are Changing the Engineer's Role

#### By Gary A. Breed Editor

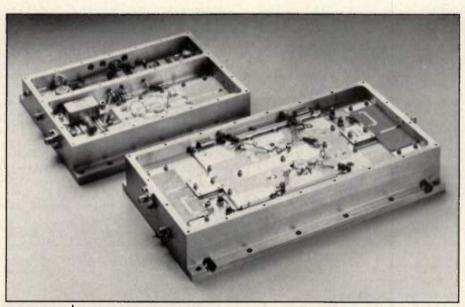
Modules, hybrids, subassemblies, board-level integration — whatever you call them, they are "building blocks" for RF systems and they are the hottest part of the RF industry right now. A look at the advertisements and new product announcements in this magazine (or any other electronics magazine) should prove that point. Although modular circuits have been around for a long time, they recently have become so widespread that engineers generally consider them as if they were individual components like resistors or capacitors.

This report is more of a news story than an update on technology. Many of the rapidly growing companies in the RF industry are in the business of making products best described as building blocks. It really doesn't matter whether their products are thick film hybrids, packaged circuit assemblies or board-level products. They represent functionally complete subsystems, and they are changing the way engineers design RF equipment.

One reason for recent growth in modular components is lack of available expertise. *RF Design* has reported the demand by RF companies for more engineers with analog design abilities (May 1987 issue, p. 57). With this current shortage of designers, companies often find themselves lacking the in-house capability to design



Oscillators are perhaps the most well known hybrid modules. This photo shows the construction of a VCXO from Reeves-Hoffman.



Known best as a transistor and hybrid amplifier maker, TRW RF Devices is now manufacturing these UHF class A amplifiers.

certain types of circuits. When a new technology comes along, the problem is multiplied.

For example, surface acoustic wave (SAW) technology has matured in just the last few years. SAW device manufacturers agree that their greatest difficulty in attracting customers is that engineers lack adequate knowledge in SAW applications. Two companies, Andersen Labs and RF Monolithics, have specifically stated that a major reason they have undertaken modular assembly is to circumvent this problem. By providing their SAW devices in an assembly containing all interfacing circuitry, they have substituted their specialized expertise for the uncertainty of the customer's own engineering staff.

Performance is another major reason for the growth in modules. As RF technology has become more complex, a natural response is to create areas of specialization. Buying a modular assembly gives one company access to the specialized ability of another manufacturer. As another example, how many companies build their own precision oscillators? Not many. Instead, they buy them from one of the many oscillator specialists. The phrase applied here is certainly, 'Why reinvent the wheel?'' In just the past two or three years, that principle has been expanded to include power amplifiers, noise sources, IF amplifiers, and entire receiver front-ends. The build-or-buy decision in RF is increasingly "buy."

#### **Cost and Competition**

Outside of consumer electronics, RF is not a mass market. The quantities of units produced by RF manufacturers are relatively small. For many years the attitude has been that quality and performance override all other concerns; that people buy equipment for those reasons. Add the desire for complete control over all design and manufacturing parameters and you have a completely engineering-driven sys-



Among the oldest modular components are filters, such as this product from TTE.

tem. For a long time, this system worked well, but there is a definite shift in economic climate that is forcing a change of attitude.

Last November, Hewlett-Packard introduced a new low cost spectrum analyzer. To create that product and sell it at a competitive price, H-P modified its historic 'engineering first" philosophy. The *marketplace* determined the price range and dictated the design goals for the product. Like another test equipment giant, Tektronix, had done a few years earlier with a new line of oscilloscopes, H-P engineers had to create an instrument within a welldefined framework of time, cost, and manufacturability.

There is another side to the cost equation: manpower. RF engineering salaries are high, and the required support equipment and facilities are expensive. As worldwide competition has grown, companies can no longer justify the cost of designing every circuit in that design. As a result, exercising control over every aspect of design and performance is more costly than ever. Fortunately, competition among module suppliers has developed to the point where quality is generally high in these system components.

#### Building Blocks Change the Designer's Role

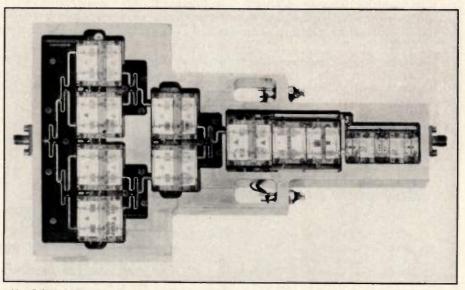
The ability to bypass component-bycomponent design of amplifiers, mixers, or oscillators changes the method of circuit design. The RF engineer needs sufficient understanding of these circuits to establish appropriate performance specifications, although he or she won't actually design them. Instead, that engineer must have top-notch ability to *integrate* various subsystems into a working whole. Formerly, the staff engineer had the opposite priorities: circuit design expertise with some system-level familiarity.

This design process is much less interactive, as it would be under the conventional arrangement where a circuit designer works under a project manager with authority over several subsystem teams. With outside vendors, the process needs to have far more advance preparation. Initially, specifications must be developed for selection of system components, including the compromises in performance that are usually necessary to meet budgets and timetables. Although the design process has always had these first steps, using outside sources for part of the system makes it very difficult to modify the system or explore different performance options as the project develops.

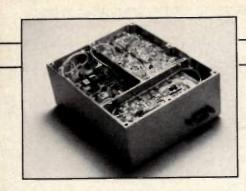
#### Product Knowledge is a Necessity

Besides the change in planning and specification of a design, modular components require that engineers at the lowest staff level be familiar with available products. If custom or semi-custom assemblies are needed, the capabilities of supplier companies should be well understood. The purchasing department may make the financial arrangements for procuring the modular components, but it is always the engineers who decide exactly what the specifications are, and which companies are most likely to be able to meet them.

This necessary awareness by engineers is in contrast with past practice. Designing with discrete components has involved "generic" resistors, capacitors, and semi-



M/A COM MPD has incorporated amplifier modules into a 20 watt S-Band power amplifier assembly.



This UHF receiver from General Services Engineering can be considered a "module" intended as part of a complete system.

conductors. While preferences for certain manufacturers' products develop as a designer gains experience, these choices do not have to made for every project. The selection of amplifier, oscillator, synthesizer or mixer subassemblies has to be done "from scratch" every time. An engineer has to be aware of the differences in performance and manufacturing expertise among several potential suppliers.

Right now, the engineers' task is most difficult. The number of companies supplying modular components is growing rapidly, requiring a continuing effort to keep up to date on products and capabilities. Also, firms which manufacture discrete components are incorporating "value added" assembly of hybrids or boards based on their discrete products. A few examples include: Motorola, which is manufacturing an RF modem board; TRW, offering a line of hybrid and boardlevel power amplifiers; Piezo Crystal Co. continues to develop new oscillator modules; and Acrian is making UHF-TV and S-Band power amplifier assemblies. All of these companies, known best for their discrete components, have added modular products to their capabilities.

#### A Gradual Rise to Prominence

"Building block" design is not new. In particular, assembled oscillators, mixers and filters have been around for a long time. What is noteworthy is that the concept has rapidly expanded to include nearly every RF function over the past few years, to the point where modules are replacing discrete designs. The modest collection of modules available only five years ago has grown dramatically, to the point where they are a factor in *all* design decisions.

There is a negative side to the phenomenon, now that it is possible to build complete RF systems by purchasing modular assemblies, interconnecting them



Rather than design an amplifier, engineers have the option of using a broadband hybrid gain block like this one from the Anzac Division of Adams-Russell.

and applying the proper DC power. There is the danger that engineers will stop *designing* and just pick assemblies from a catalog and put them together in a system. RF engineers have to realize that the process of design should not be less rigorous because modular products are used.

Like all new developments in product technology, the availability of these supercomponents offers outstanding time and convenience benefits. With proper system planning, maximum performance can also be achieved. But, also like any new development, users of the technology must understand the changes it brings to the process of engineering.

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## rf featured technology

## **Specifying High Power Filters**

Important Considerations for Filter Design and Purchase Decisions.

#### By Dick Wainwright QRS

There are still plenty of engineers who say, "Anyone can design filters, the textbooks are full of tables of element values." That is true, but there are many more pertinent specifications than just L and C values. Component ratings, not the values, are of fundamental importance in power handling devices. The old saying, "A little knowledge is dangerous," applies to high power filter design — resulting in a lot of smoke testing.

Take a moment to ponder the long list of key terms in the language of high power filter design. Electrical considerations include: Power, Q, selectivity, energy storage, modulation, peak-to-average, harmonic content, rejection, EMI, topology, dielectric strength, losses, and so on. Equally important are the mechanical and environmental aspects: heat generation, flow and sinking, contaminants, vibration, salt spray, connectors, altitude, plating, and many others. These terms are but a smattering of the flow of words that haunt every sensible designer of high power devices.

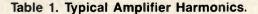
It takes an intimate knowledge and an awareness bordering on paranoia, along with considerable experience, to tackle this area of design. The same understanding is needed when specifying a filter for purchase from an outside vendor. More than one manufacturer has experienced the frustration of losing a bid because of exceptions prudently taken to allow for poorly defined specifications. Many times the user does not know or furnish such important information as:

a) *Harmonic content* of transmitter power output relative to fundamental power.

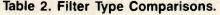
b) *Connectors* which may be incompatible with specified power and load conditions.

c) Surface area available for heat conduction/radiation and availability of cooling air. Assuming that cooling air is available, its rate of flow, pressure and temperature.

Harmonic Order	Level (dBc)	Harmonic Power W½kW (Fundamental)
2nd	-18	15.8
3rd	-12	63.1
4th	-21	7.9
5th	-18	15.8
5th	-22	6.3
7th	-21	7.9
(8-13th)		37.8 (total)



Selectivity 60 dB/3 dB	Nominal VSWR	Filter Design	Qul for a Given Loss	Energy Storage
1.25	1.2	Elliptic (0.01 dB ripple)	1000	20
1.5	1.35	Chebyshev (0.1 dB ripple)	650	15
1.7	1.2	Chebyshev (0.01 dB ripple)	460	12
-	- 54	Chebyshev (0.001 dB ripple)	290	-
2.1	1.2	Butterworth	200	10
4.2	-	Bessel	-	



d) Proper shock and vibration specifications. A vivid recollection concerns heavy shock and vibration specs along all three axes for an application which, under power, was fixed and indoors!

e) VSWR with realistic values. 1.1:1 is not the right specification for a filter which would, in practice, be operating into a load of 2:1 or 3:1, at almost any phase angle. f) Other mitigating relationships which combine to specify an unrealistic design.

#### **Harmonic Content**

Typical solid state transmitters of recent vintage can be expected to yield harmonic power levels such as those listed in Table 1. Besides the filter requirements, the system designer may opt to use a ferrite iso-

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lator at the transmitter output to maintain a well-matched condition. However, most isolators are frequency sensitive, and the result may be excessive heating due to the harmonic energy. Burnout is one possible consequence, as is additional harmonic generation from operation of the ferrite material near its Curie temperature.

#### **Connector Ratings**

Most connector manufacturers don't know how much power or apparent power their connectors will safely handle under all possible operating conditions. Current ratings rather than voltage ratings are generally the problem area.  $Z_o$  is of little importance in systems working with very high VSWR values (although, of course, they are related).

The fact is that most filters are reflective devices, producing attenuation by reflection of power back to the source, rather than by absorption. Absorptive filters are generally more expensive and considerably larger than reflective types. However, especially at UHF and higher frequencies, the use of harmonic absorptive filters should be considered.

#### **Cooling Considerations**

How much surface area should one allow for the cooling of high power filters? If cooling air is available, what is its pressure and rate of flow? First, we make the following assumptions: 1. The maximum power versus connector type is in the order of 2/3 of the rating of the cable attached, derated appropriately; and 2. A high power filter is normally designed such that the hot surface temperature does not exceed ambient plus 50C, with a heat sinking plate average temperature that is no more than ambient plus 10C (95C maximum). Cooling air, if available, should not exceed an effective temperature of 80C; hence, air speed may be an important consideration as the air approaches the filter hot spots.

In many cases, substantial size fins may be required, but fins are *not* effective in cramped air flow spaces. Convection air currents must be free to circulate if convection represents a substantial portion of the cooling means. Surrounding heat conducting surfaces may need to have rough textures to "wipe" the heat out of the circulating air.

#### **Shock and Vibration**

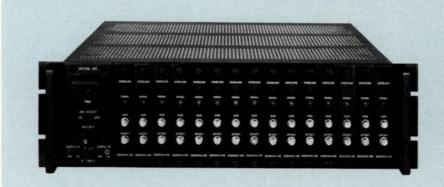
Heavy shock and vibration specifications along all three axes should be studied carefully. The strong G forces should be confined, if at all possible, to the smaller dimensions of the coaxial structure. Minimal forces of no more than 5G (preferred), or 10G maximum (11 m/s standard shock specification) should be applied on the length of coaxial filters of substantial size. In addition, the adequacy of the mounting and supporting hardware is essential.

#### **Electrical Specification**

Selectivity, time delay, energy storage, electrical (absorptive) losses and heat

generation go hand-in-hand. Q, the ratio of *Energy Stored per Cycle* to *Energy Dissipated per Cycle* of the applied energy is a critical consideraton in the design of all filters in general, and of great importance in high power filters. For a given loss, the required ratio of unloaded Q/loaded Q is different for every filter design. It is not uncommon for certain designs to require inductor unloaded Q val-

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ues in excess of 500-1000 and for capacitors, 3500-8000. Even with these values. the engineer may only eke out a responsive design where rejection bandwidths to low loss band edge ratios are small, e.g. much less than 1.2:1 for 40-60 dB rejection. The almost impossible Q values of 2000 or more for coils and over 10,000-15,000 for capacitors may be required, but are unobtainable in practice because of space, moding, or frequency limitations.

The Q of coils increases roughly as the square root of frequency, so high frequency filters generally yield lower losses for the same selectivity and available space.

For a given loss (see Table 2) the unloaded Q required increases substantially with passband ripple (VSWR), complexity (n) and type of filter (Elliptic, all-pole Chebyshev, Butterworth, etc.). Note that Table 2 indicates representative values, not absolutes.

As an example, a 0.01 dB Chebyshev (all-pole) filter required approximately (460/200-1), or 84 percent more Q than a Butterworth filter for a given loss. The Elliptic filter given has twice the energy storage indicating that voltages and currents are roughly 40 percent more than in a Butterworth filter.

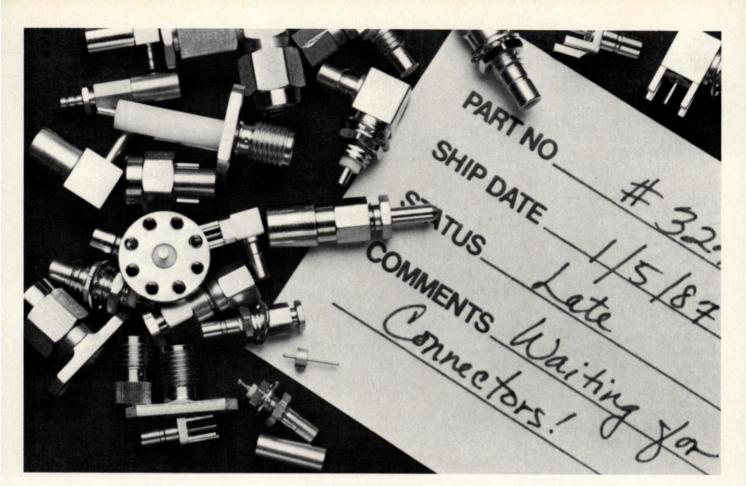
#### **Other Factors**

There are several factors that also require consideration in the specification and design of high power filters. These include phase linearity versus frequency in a single filter, or phase matching in filter groups. A ± 5 degree phase matching specification may double the price relative to a non-matching phase unit, to allow for extensive alignment and component selection. Additional work may be required in mechanical construction to assure matching under varying operating conditions.

Humidity is another concern. High humidity, especially when combined with condensation, can present substantial difficulties. Another environmental concern is altitude of operation. Derating may be necessary, and in difficult cases it may be best to pressurize the unit with dry nitrogen or sulfur hexafluoride. Of course, when pressurized, humidity ceases to be a significant problem. rf

#### About the Author

Dick Wainwright has many years experience in filter design as founder and Chief Scientist of Cir-Q-Tel. He is now practicing as a consultant. He can be reached at QRS, PO. Box 116, Kensington, MD 20895.



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# **rf** design feature

# **Phase Noise Calibrator**

#### By Guy V. Love and Kevin Lindell Norden Systems

The spectral purity of RF signal sources used in today's radar and communication systems is often the limiting factor in achieving the desired performance. Such spectral purity is usually specified in terms of phase noise, a measurement of the spectral distribution of the random (or regular) wanderings in phase of the ideal sinusoid, by which the source is characterized. The authors earned the runnerup prize of EEsof's Touchstone/RF in the second annual RF Design Awards Contest.

Source manufacturers often assemble their own test sets for measuring phase noise using a few items of standard test equipment, an appropriate source and a few in-house circuits which vary depending on the technique of choice.

In the procurement of low phase noise sources there is a need to assure that measurements made at the vendor's facility, for instance, on test set "A", can be duplicated in the buyer's lab on test set "B". A calibrator that can be used with any type of test set, whose performance is reliable and easily verified, and which is easily transported, has been developed. Its description, details of construction and application follow.

#### Technique

The basic idea, shown in Figure 1, is to introduce a general type of phase modulator ahead of the phase noise test set. The RF output of the source is switched between two lengths of coax, one of which is slightly longer than the other. This small change in electrical length appears to phase shift the source-under-test at the clock rate, producing a known phase modulation. The spectrum produced is also known, and provides the means of calibrating the phase noise test set.

The calibrator can be turned off to inhibit interference with the noise measurement of the source. The technique is inherently low loss and, being coaxial, is well shielded to assure negligible additive noise or interference. The approach is broadband, and one set of transmission lines can be used over several octaves. The phase deviation produced by switching between the two lengths of coax is related to the difference in coax lengths,  $\Delta L$ , by the following equation:

$$\Delta \phi(pp)$$
 in radians =  $2\pi f \frac{\Delta L}{V_{c}}$ 

where f = frequency of source under test

- $\Delta L$  = difference in length of cables
- V<sub>p</sub> = phase velocity of coax cable used, at frequency f.

The phase shift is frequency sensitive varying linearly with frequency as long as phase velocity is constant. It is desirable to keep the phase modulation small so that the spectrum can be predicted from small index approximations and also to provide a calibration commensurate with the noise to be measured. At the maximum frequency of usefulness, determined by the transmission band of RF switches used, the coaxial cables should differ by less than 0.1 radian if small index theory is used.

#### Operation

In general, and specifically for large phase deviations, angle modulation is a non-linear process, producing an RF spectrum whose amplitudes are in terms of Bessel functions. However, if the modulation index is kept small (<0.1 radian), the following approximations to these functions can be made:

- $J_0(\Theta) = \text{carrier amplitude} \cong 1$  $J_1(\Theta) = \text{amplitude of first order}$ sidebands  $\cong \Theta/2$
- and  $J_2(\Theta) = J_3(\Theta) = J_4(\Theta)$ , etc.  $\cong 0$

where  $\Theta$  is the modulation index, which is equal to the peak phase deviation in radians, and J<sub>n</sub> represent Bessel functions of the first kind and order n. Only the first order sidebands are significant, and they are related in amplitude to the carrier by

#### $dBc = 20 \log \theta/2$

For example, if single frequency sinusoidal phase modulation is used, with 0.1 radians peak-to-peak:  $\Theta$  = modulation index = 0.05 radians peak, and 20 log  $\Theta/2$  = -32 dBc. In the spectrum, single lines of this value would appear, spaced either side of the carrier by the modulation frequency.

The small angle approximation also permits superposition to be used to predict the spectrum produced by more complex modulation waveforms. Each of

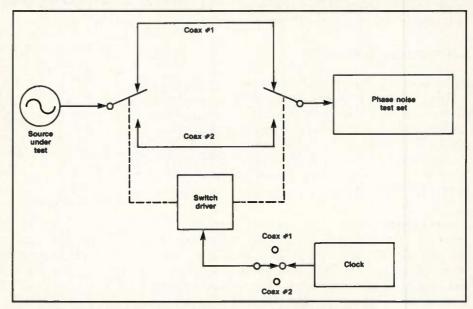


Figure 1. Phase noise calibrator — principle of operation.



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the Fourier components of the modulating signal then produces its own first order sideband pair around the carrier. For the case of a symmetrical square wave, only odd harmonics exist. The fundamental component (at  $f_m$ ) produces a pair of lines  $\pm f_m$  about the carrier; the third harmonic produces a pair  $\pm 3f_m$  about the carrier, and so on for 5th, 7th, etc. The

calibrator uses such a symmetrical square wave phase modulation, switching between the two coax cables whose phase difference is  $\Delta \phi$ (pp). The Fourier components of the modulation function are given by:

~ / >	∆¢(pp)		for n=1, 3, 5
⊖(n) =	2	πη	and zero for
	-		n=even

The phase modulation sidebands are then given by:

$$\frac{\Theta(n)}{2} = \frac{\Delta \phi(pp)}{\pi n} \text{ for } n = 1, 3, 5, \dots$$

And, relative to the carrier level C, the single sideband levels L(n) are:

$$L(n) dBc = 20 \log \frac{\Theta(n)}{2} =$$

$$20 \log \frac{\Delta\Theta(pp)}{\pi n} \text{ for } n = 1, 3, 5, ...$$

For the case  $\Delta \phi(pp) = 0.1$  radians, the following sidebands are given

L(1) = 20 log  $\frac{0.1}{\pi}$  = -29.9 dBc L(3) = 20 log  $\frac{0.1}{3\pi}$  = -39.5 dBc

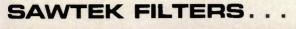
L(5) = 20 log 
$$\frac{0.1}{5\pi}$$
 = -43.9 dBc, etc.

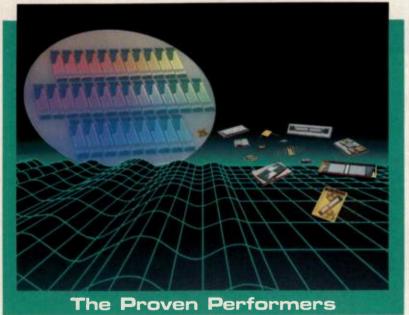
As expected, the sidebands fall as 1/n, equivalent to -6 dB/octave on a log-log plot of phase noise. The peak values of these line spectra are used to calibrate the plotter output when measuring the phase noise of an unknown source. The choice of clock frequency (fundamental) is not a factor in calibration, and is largely a matter of convenience. It should be high enough to avoid being demodulated by the phase locked loop commonly found in phase noise test sets. It should not coincide with line frequencies or their harmonics, which are often large. An acceptable value is a 100 Hz clock.

#### **Calibration And Sources Of Error**

In theory, a careful measurement of the difference in cable lengths, together with an accurate value for cable phase velocity, would yield a close value for  $\Delta\phi(pp)$ . However, added differential phase shifts in the RF switches, slight but invariable differences in construction, and the uncertainty of the phase velocity, all contribute to errors in this approach. A direct measurement of  $\Delta\phi(pp)$  is possible using a network analyzer, or a vector voltmeter.

For use as calibrator, it is the spectral line amplitudes relative to the carrier (dBc) that must be known. As was shown, these vary directly with  $\Delta\phi$ . A 0.5 dB accuracy in spectrum magnitude requires better





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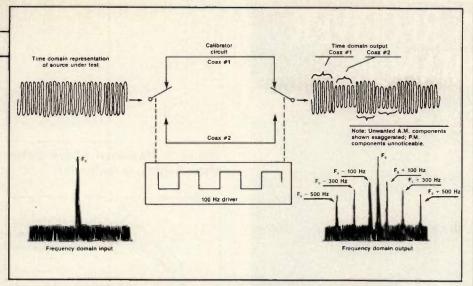


Figure 2. Signal representation in the phase noise calibrator.

than 6 percent accuracy in measuring  $\Delta\phi$ . If  $\Delta \phi$  is 6 degrees, the measurement error should not exceed 0.36 degrees. Since  $\Delta \phi$  is the difference of two larger phase shifts, the allowable percent error, in terms of either path length, must actually be much less than 6 percent. Fortunately, absolute phase shift is not needed, and the evaluation of A allows systematic or offset type errors to be differenced out. Still, 0.5 dB accuracy requires considerable

care. A 1.0 dB uncertainty in spectrum content is more easily achieved and may be adequate for many phase noise measurements.

An experimental model of the calibrator was built, and  $\Delta \phi(pp)$  evaluated using a network analyzer. The transmission phase angle was measured for each path, using the mode switch to select the path. Included in the path is any phase shift due to the RF switches. It is essential that these



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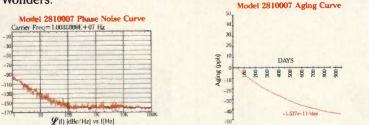
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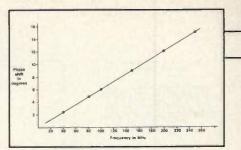


Figure 3. Measured phase difference,  $\Delta\phi(pp)$ , in calibrator.

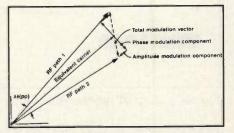


Figure 4. Vector representation of small index phase modulation plus amplitude modulation.

switches exhibit the same phase delay when switched statically as when driven by the clock. Since their phase shift was found to depend on drive level, the RF switches are always controlled by the same drive circuit (a TTL switch driver), the DC voltage (+5 V) is regulated, and mode switching is done ahead of the driver.

While the calibrator was intended for use in the 50 to 200 MHz range, the Ad value was measured over a somewhat wider range. No dispersion (variation in phase velocity with fequency) was seen, and the data points fell closely in a line, as shown in Figure 3. A best fit straight line averaged the data and facilitated interpolation of values of A at different frequencies. The measured value of  $\Delta\phi(pp)$ was found to be 6.1 degrees at 100 MHz, and the first order sidebands were expected to be -29.4 dBc. As a check, the modulated output of the calibrator was observed on a spectrum analyzer. All of the sidebands were greater than expected, with the first pair at -28 dBc. In looking for the discrepancy (about 1.4 dB), it was realized that the spectrum analyzer does not distinguish between amplitude and phase modulation, but measures the spectral energy distribution due to all causes. If AM was present, it could explain the increased sideband levels.

Figure 5 shows how the AM and PM vector components add to produce an increase in the total modulation content. The amplitude difference in the vectors (labeled RF path 1 and RF path 2) represents a difference in the insertion loss through the two paths (including switch losses), while the phase difference in these vectors is the measured phase

modulation. Looking again at the network analyzer results, an amplitude difference of about 0.5 dB in the path losses was measured. The fractional amplitude difference, corresponding to  $\pm 0.25$  dB, would be  $\pm 0.03$  relative to a unit carrier vector. The measured phase modulation of 6.1 degrees pp translates to be  $\pm 0.053$ radians. The total modulation vector is the quadrature addition:

$$\sqrt{(0.03)^2 + (0.053)^2} = 0.061$$

The total peak modulation vector, 0.061, represents about 1.2 dB increase over the expected 0.053 vector for phase modulation alone. This explains most of the 1.4 dB discrepancy in sideband energy as measured with the spectrum analyzer. The remainder is easily within the expected error of measurement. Confidence that the phase modulation component is close to the value that was expected, nominally -29.4 dBc, is important because a phase noise instrument is sensitive only to phase modulation (assuming proper alignment) and is insensitive to AM. An accurate measure of phase angle  $\Delta \phi(pp)$  throughout the band, therefore, should be sufficient to establish a useful calibrator.

#### Construction

Figure 5 is a schematic diagram of that model. IC1 is a clock generator, set to produce four times the desired modulating frequency. IC2 is a dual D flip flop which divides the clock frequency by four, assuring a 50 percent duty cycle square wave with which to drive the RF switches. IC2-b has its *Set* and *Reset* lines tied to a three position manual switch to permit selection of the toggling mode, or either coax line #1 or #2. This enables the coax lines and switch combinations to be characterized on a network analyzer. Also, the calibrator can be left in the test set circuit and disabled to permit phase noise measurements to proceed without calibrator interference.

SW#1 and SW#2 are pin diode type RF SPDT switches, whose passband largely determines the RF range and characteristics of the calibrator. The switches should have a reasonably fast switching time (a few orders of magnitude less than the modulating period) to assure nearly ideal modulation. They should also be closely matched in path loss, together with their coax lines, to prevent excessive amplitude modulation. Coax #1 and #2, constructed of semi-rigid (0.141 OD) 50 ohm coax, are of different electrical length to produce a small phase shift in the frequency range of interest. The switches used were fitted with SMA connectors to allow different coax lengths to be easily accommodated.

The measured value of  $\Delta \phi(pp)$  was 6.1 degrees at 100 MHz, as has been mentioned. Using a 100 Hz switching rate, the expected nth order sideband levels are given by:



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Figure 5. Schematic diagram of experimental calibrator.

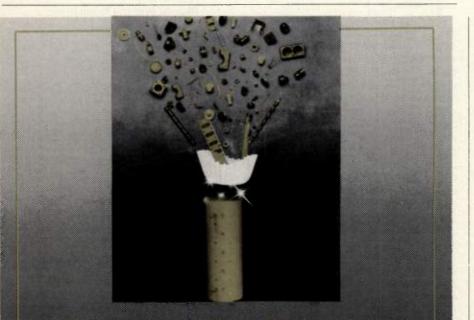


and are:

-29.4	dBc	at	100	Hz	(n	=	1)	
-38.9	dBc	at	300	Hz	(n	=	3)	
-43.4	dBc	at	500	Hz	(n	-	5),	etc.

#### Applications

The calibrator can be inserted after either of the RF sources in a two-source type phase noise test set. Figure 6 shows the calibrator in series with the voltage controlled reference source (usually an internal component or synthesizer). The modulation rate (clock frequency) must be well above the phase-locked loop bandwidth, or the loop will act to demodulate the phase shift, reducing the spectral content and upsetting the calibration. This is true whether the calibrator is in-



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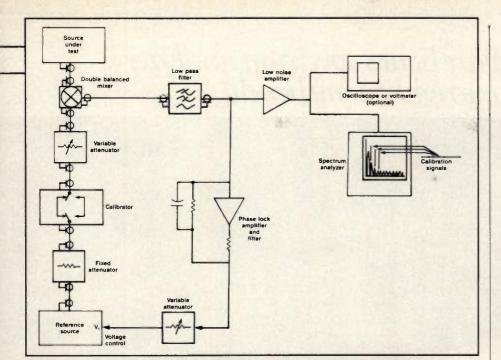
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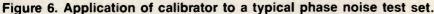
side or outside the loop. In a discriminator-type phase noise instrument there is only one source (the unknown), and, without a locked loop, there is no such limitation on clock frequency.

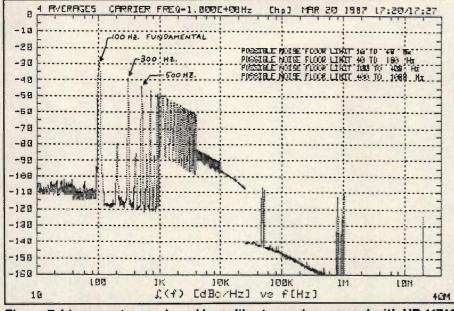
The calibrator was used to check an HP 11740 phase noise test set, the plotter output of which is shown in Figure 7. A 100 MHz source was used, and the expected calibration spectrum was -29.4 dBc for the first order line at 100 Hz offset, -38.9 dBc at 300 Hz offset, and -43.4 dBc at 500 Hz. Phase noise is defined in terms of the power density on one side (single sideband) of the carrier, so the above line levels should not be doubled (increased by 6 dB) to account for the double sideband nature of the RF spectrum.

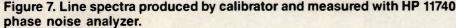
As can be seen in Figure 7, agreement with expected values is quite close. The small scale of the plotter output precludes resolution better than 1 dB, and the accuracy goal of ±0.5 dB for the calibrator appears to be entirely adequate for the purpose. The higher order sidebands continue to fall 6 dB per octave, as is expected. The closeness of the lines on a log plot become so great beyond 1 kHz that the plotting routine never gets down to the source noise floor between lines. Ultimately, the analyzing bandwidth is inadequate and the plotter output beyond 4 kHz is meaningless. A possible source of misunderstanding is whether the calibration levels should be adjusted by 3 dB. just as it is assumed that a noise plot can be reduced 3 dB to account for equally noisy sources, each of which contribute one half of the observed phase noise. No such adjustment of the calibrator level is needed.

It is helpful to realize that the HP 11740 reads out total phase noise, including that of the reference source and the unknown source, making no assumption about relative noisiness. This requires any appropriate reduction in the plotted noise level to be made by the operator, depending upon his knowledge of the noisiness of the source-under-test. If the source-undertest is known to be much noisier than the reference, it dominates the measurement and no reduction in measured value of phase noise should be applied. Similarly, since only one of the sources is being modulated for calibration, no adjustment should be made to the plotted value. The number -29.4 dBc, for the first-ordered sideband is an absolute reference point for the detected phase noise, whether it arises because only one source is noisy or because both are noisy. In the latter case, only the noise plot should be adjusted downward to report the single









source phase noise, and not the calibration of the plot.

#### Summary

The calibration of phase noise measuring instruments can be established or confirmed using a simple phase modulator, built around the principle of switching between two short sections of coaxial cable. Such a modulator would be placed in series with a source to be measured, and, provided the line difference is less than 0.1 radian at the source frequency, the line spectra which result have a usefully small and easily determined value relative to the carrier. The amplitude of the line spectra can be used to establish the calibration of a phase noise test set to within  $\pm 1$  dB. The technique can be applied over the entire frequency spectrum, subject to the availability of suitable components. Multiple calibrators can be cascaded to provide coverage of a wide frequency range with a single instrument. Norden Systems has filed for patent disclosure on this technique.

#### About the Authors

Guy Love is a senior design engineer and Kevin Lindell is an RF engineer at Norden Systems, Inc., Division of UTC, Norwalk, CT 06856. The authors can be reached at (203) 852-5000 for Guy, and (203) 852-5148 for Kevin.

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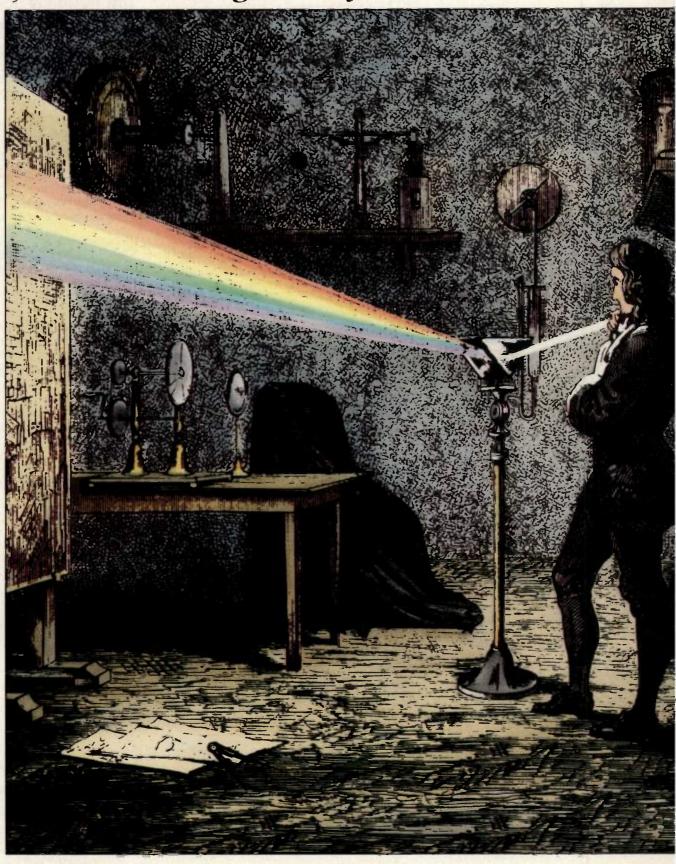
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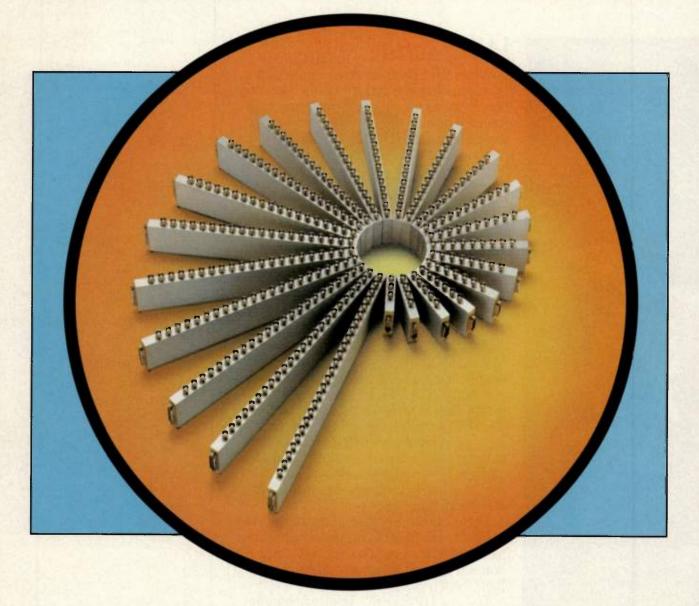
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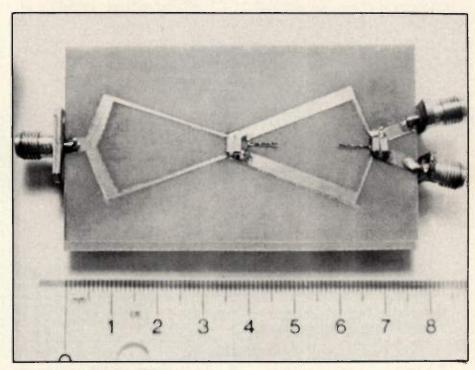
### Space-Saving Microstrip Design is Contest Runner-Up

#### By Peter Vizmuller Motorola, Ltd.

The frequency range between 500 MHz and 2 GHz presents a challenge for broadband power splitters. Those based on ferrite transformers become difficult to optimize above 500 MHz and broadband power splitters based on conventional microstrip techniques tend to be bulky below 2 GHz. The design in this article illustrates an inexpensive, compact microstrip power splitter that covers an octave bandwidth, yet is smaller than conventional microstrip realization. This design won Peter Vizmuller the Bird Model 4421 RF Power Meter in the RF Design Awards Contest.

The three most important design parameters for a power splitter are input match, output match and isolation. Other performance characteristics, such as amplitude and phase imbalance depend on circuit symmetry and are determined by construction techniques. The input match problem can be reduced to a 100 ohm to 50 ohm impedance transformation, as shown in Figure 1.

In a 50 ohm system, an equal in phase power split will occur if each of the output lines presents a 100 ohm load to the input line at A-A shown in Figure 1. This way the input line will see two 100 ohm loads in parallel, and a 50 ohm input match will be maintained. Each output line has to behave as a 100 ohm to 50 ohm transformer. The simplest impedance transformer is a quarter-wave length of 70.7 ohm transmission line, which forms the basis of the classical Wilkinson power splitter (1). The disadvantage of this impedance matching method is its narrow bandwidth. More sections of quarter-wave lines of varying characteristic impedances improve the bandwidth but make the transformer much longer.



The miniature power splitter.

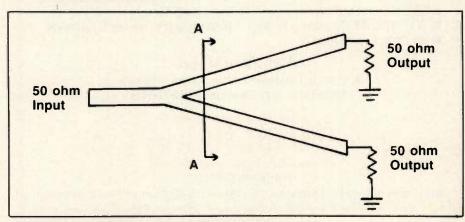


Figure 1. Simplified power splitter.

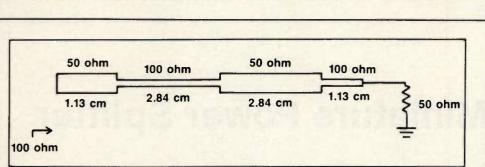


Figure 2. 100 ohm to 50 ohm nonsynchronous transformer.

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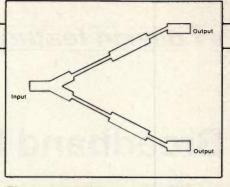
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#### Figure 3. 50 ohm power splitter not corrected for output match and isolation.

Another approach to this impedancematch problem is the use of nonsynchronous transformers(2), where in matching 50 ohm to 100 ohm source and load impedances, sections of 100 ohm and 50 ohm transmission lines can be used to accomplish the impedance transformation. This is the approach taken for the proposed splitter.

A circuit analysis program was used to design a nonsynchronous transformer to operate between 450 MHz and 950 MHz on glass-epoxy circuit board as shown in Figure 2. The total length of this transformer is equal to the 70.7 ohm singlesection tranformer, but it covers almost three times the bandwidth. Having obtained the required input match, the next step is to look at output match and isolation. Figure 3 shows the uncorrected power splitter. This splitter provides an equal power split from the input to the output in a 50 ohm system, but the output match and isolation is poor. To improve both, an isolation resistor can be added between the two outputs as shown in Figure 4.

The resistor isolation is accomplished the following way: Assume a signal is fed into one output, terminating the input in 50 ohms and looking at power coming out the other output. The signal that travels from one output to the other via the transmission lines is delayed by 180 degrees due to the quarter-wave length

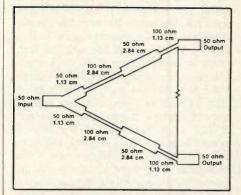


Figure 4. Power splitter with one isolation resistor.

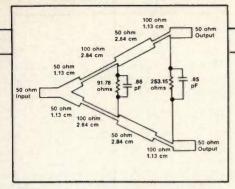


Figure 5. Final splitter configuration.

of each transformer, but the signal which takes the path through the resistor is not delayed, and therefore out of phase with the other signal. Selecting a proper value resistor ensures equal amplitudes for the two signals and therefore signal cancellation at the other output, accomplishing isolation.

Using one resistor achieves isolation over a narrow bandwidth, comparable to a conventional Wilkinson splitter. To increase the isolation bandwidth, two resistors are used as shown in Figure 5. The other resistor improves isolation at higher frequencies, since some of the signal going to the input is shunted to the other side by this resistor, decreasing the phase shift as required at higher frequencies, where the impedance transformers are longer than quarter-wave and would provide more than 180 degrees phase shift without the second resistor.

The splitter as constructed so far, has good input match, output match and isolation, but these parameters do not coincide exactly in frequency. Small capacitors added in parallel to the isolation resistors overcome this difficulty by shifting the output match and isolation curves in frequency, so that all three parameters become centered around a common frequency. Note that any components placed between the two output lines at corresponding points do not affect the input match or power split, since

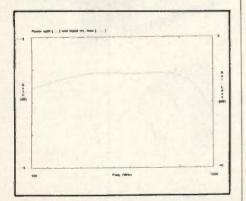


Figure 6. Theoretical performance of the splitter.

signals at corresponding points in the output lines are in phase (when the input is the signal source), and there is no potential difference across a component placed there.

The final resistor and capacitor values were obtained by a circuit optimization program, design criteria was better than 20 dB input and output return loss and better than 20 dB isolation over 450 MHz to 950 MHz. Figures 6 and 7 show the theoretical performance. The power splitter was then constructed on standard G-10 glass epoxy board, 1.6 mm thick. The transmission lines were folded at their junctions to allow the use of chip isolation resistors; 50 ohm lines were 2.7 mm wide and 100 ohm lines were 0.54 mm wide. Capacitors were realized as short lengths of twisted bifilar wire. Measured performance of this splitter is shown in Figures 8 to 12.



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Device	Freq. (GHz)	P,dB (dBm)	G,dB (dB)
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FLL171ME	2.3	32.5	12.5
FLL351ME	2.3	35.5	11.5
FLL50ME	2.3	36.0	10.0
FLL100ME	2.3	39.0	8.0
FLL2001B-2	2.3	42.5	11.0



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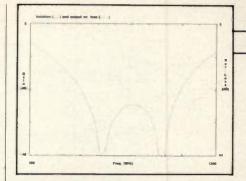


Figure 7. Theoretical performance of the splitter.



Figure 8. Measured input return loss.

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#### 50R-083

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#### 50R-084

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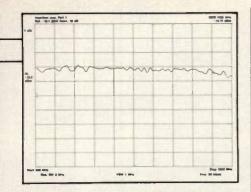


Figure 9. Insertion loss at Port 1.

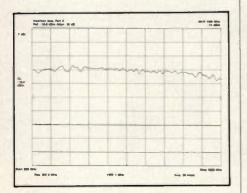


Figure 10. Insertion loss at Port 2.

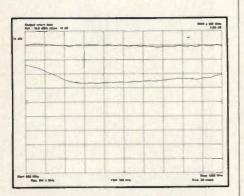


Figure 11. Output return loss.

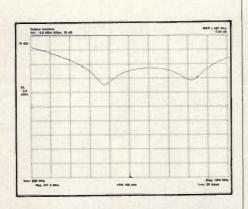


Figure 12. Measured output isolation.

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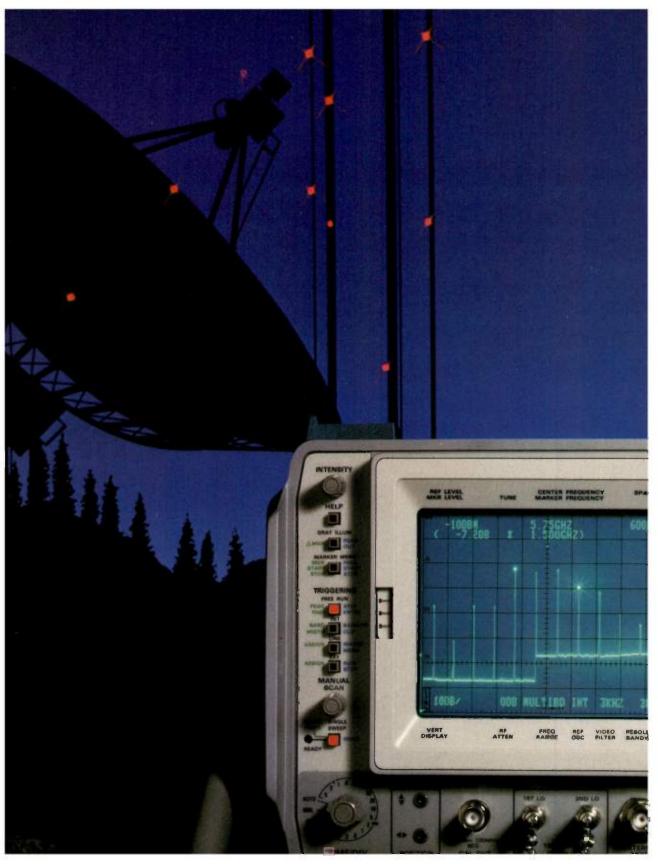
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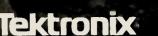
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#### Results

The design objectives of 20 dB return loss and isolation were achieved from 460 MHz to 1030 MHz, bandwidth performance being limited by the input return loss; output return loss and isolation covers a wider bandwidth. Maximum insertion loss measured was 3.4 dB at the higher frequencies, maximum amplitude imbalance was 0.2 dB. The measured prototype is shifted higher in frequency from the theoretical design, probably because in folding the lines at right angles, the high impedance lines are slightly foreshortened.

Table 1 outlines the performance of a Wilkinson splitter, the author's power splitter and a conventional two-section microstrip splitter of comparable performance (3), all realized on glass-epoxy circuit board, covering the same frequency band (results were obtained by computer simulation). The summary table shows

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70.7 ohm 7.64 cm Wilkinson: 100 ohms 70.7 ohn 7.94 cm 52 ohm 5.7 cm Conventional 83 ohm 5.7 cm 233 105 ohms ohms 83 ohm 5.7 cm

Solitter

### Table 1. Comparison between three splitters.

62 ohm 5.7 cm

that the proposed power splitter is more broadband when compared to a Wilkinson of the same length and is 30 percent shorter than a conventional two-section microstrip version of slightly better performance.

#### Conclusion

This design illustrates how a fresh look at impedance matching coupled with understanding of basic circuit performance can advance the state-of-the-art in microstrip power splitters, to produce an inexpensive, practical broadband device.

#### References

1. Howe, H., Stripline Circuit Design, Dedham: Artech House Inc., 1974, p.94.

2. Matthaei, G., L., L. Young, E.M.T. Jones, *Microwave Filters*, *Impedance Matching Networks*, and *Coupling Structures*, Dedham: Artech House Inc., 1980, p.334.

3. Howe, H. op. cit., p. 105.

#### About the Author

Peter Vizmuller is a staff engineer at Motorola Canada, 3125 Steeles Ave. E., North York, Ont. M2H 2H6, Canada; Tel. (416) 499-1441 ext. 3693.

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Proposed:	7.94	570	2.39	OUTPUT WAVEFORM for any INPUT
Conventional: Proposed:	11.40	610 100 ohm 50 ohm 1.13 cm 2.84 cm	2.49	• 98 NODES     • COMPATIBLE with HP DATA FILES     • TIME DOMAIN     • GROUP DELAY     • RELATIVE PHASE     • NEGATIVE COMPONENTS     • NEGATIVE COMPONENTS     • ALSO: FFT WAVEFORM ANALYSIS     • A TIMES FASTER than HP     • INTEGRATED into ONE FILE
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# **Conductive Plastic Composites Offer Shielding Solution**

#### By Mark Gomez Assistant Editor

Electromagnetic interference is an unavoidable aspect of design engineering. The search for cost effective methods of shielding has brought about various solutions to reduce interference problems. Conductive plastics offer one possible answer to this dilemma. However, plastics can only be used for new designs, while other products may be used to fix or improve existing systems. This article takes a look at a plastic shielding material developed by GE Plastics.

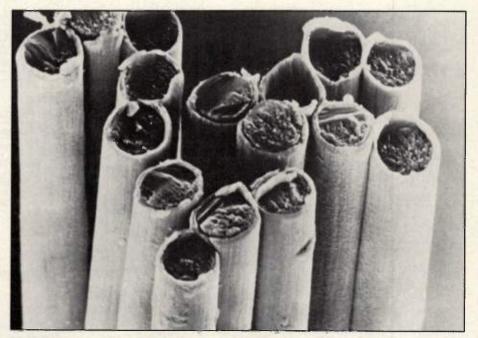
Shielding is defined as the ability of a material to reduce the transmission of propagating electromagnetic fields. It is continually being improved with the development of various fabrics, composite materials, and plastics. Conductive plastic composites are made up of polymer resins that contain electrically conductive additives.

Metal coatings such as metal filled paints and electroless plating have not proved to be effective since they are prone to flaking. Also, aluminum flake and stainless steel fibers used in shielding are relatively expensive and difficult to process. Plastics offer molded-in functionality and usually look better and weigh less than conventional methods.

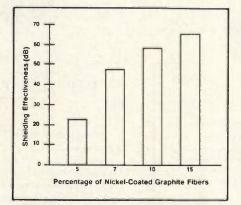
High performance resins are easy to process and provide uniform and consistent inherent shielding without the need for costly secondary operations. Plastic materials which exhibit equivalant or superior shielding are generally more economical than nickel acrylic painting, electroplating, vacuum metalization or zinc arc spraying.

The Lexan® dB5000 resin series is a thermoplastic introduced by General Electric Plastics designed to provide shielding from EMI and RFI. The Lexan conductive polymer is a combination of bundles of nickel coated graphite fibers and pellets of Lexan resin. The bundles comprise of up to 12,000 quarter-inch-long fibers that become evenly distributed during the melt process, providing the desired shielding level.

A key advantage of the resin is the molded-in shielding for increased design



Nickel coated graphite fiberts are combined with pellets of Lexan resin to achieve the desired shielding level.



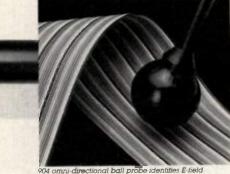
# Table 1. Effects of nickel coated graphite fibers on shielding effect-iveness.

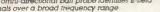
freedom. Another advantage is that the mechanical and logistic problems associated with line of sight coatings are eliminated. Since the shielding is molded in, it will not flake or peel off. Varying the loading levels in the Lexan resin/nickel-coated graphite fiber mixture gives the designer greater control over shielding levels, conductivity and cost. For example, a mixture of seven percent nickel-coated graphite fibers in Lexan resin has a shielding effectiveness of about 47 dB, which is greater than a coating of nickel acrylic spray. A 10 percent loading level of nickel-coated graphite fibers increases shielding effectiveness to 58 dB. Table 1 illustrates shielding effectiveness with respect to loading level of nickel-coated graphite fibers.

This method of shielding is fairly cost effective. One manufacturer has priced several shielding methods for an oscilloscope and has determined that at seven percent loading, nickel coated graphite fibers are less costly and provide better shielding than nickel acrylic paint.

For more information from GE Plastics on Lexan dB5000, please circle INFO/CARD #101.

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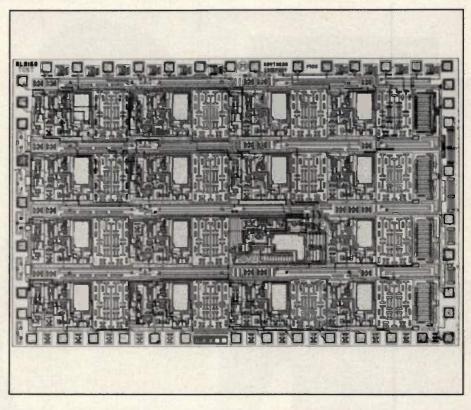


#### **Raytheon Offers A Linear Macrocell Array**

The RLA160 is a flexible VSLI linear macrocell array that provides the user with the versatility of predesigned programmable analog cells which can be converted into a variety of analog functions such as current sources, voltage references, detector/amplifier circuits, voltage to current sources, data conversion circuits, active filters and various others. With complete isolation, the analog functions can be formed from an interconnect pattern determined by the user. It can replace a circuit board with as many as 15 analog devices of 10 transistors each, 240 binarily-weighted thin film resistors, 10 PNP transistors, 43 NPN small signal transistors, four NPN 200 mA switching transistors and a bandgap voltage reference.

The thin film resistors are made by sputtering silicon-chromium onto the chip. All 240 resistors can track to an absolute temperature coefficient of 5 ppm with a typical tolerance of  $\pm$  5 percent each. All the resistors and transistors have double terminations permitting utilization of all the ICs macrocells and up to 80 percent of all individual transistors.

At unity gain, the slew rate is measured at 0.7 V/us and the bandwidth is 2.0 MHz. The input offset voltage is 1.0 mV while the input offset current is 1.0 nA. All the macrocells can be programmed for specific quiescent currents and AC operating conditions. Also sufficient gain can be obtained at specific frequencies which is a



useful feature in the design of filters.

The price of the RLA160 is dependent on the package type, temperature range, cell utilization, and custom test requirements for specific applications. Kits are available from Raytheon to aid a prospective user in developing prototypes and are intended for designers who want to make a trial application of the RLA160 macrocells in an existing system. **Raytheon Company, Semiconductor Division, Mountain View, CA. INFO/CARD #220.** 

#### 750 MHz IF Limiter Discriminator Introduced By RHG Electronics

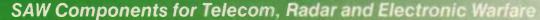
The Model ICDX750 is a delay line IF limiter discriminator that operates at 750 MHz with a peak to peak bandwidth of 450 MHz while the minimum linear bandwidth is 250 MHz. The ICDX750's wide bandwidth and 20 nsec rise time enables it to handle FM demodulation of high speed pulsed carriers where substantial AM is present. It can also be used to measure the frequency of a pulsed carrier. The unit is rated for an input level of 0 dBm, but is usable down to -10 dBm. A delay line for demodulation, coupled with a constant phase limiter/driver for minimum degradation of pulsed waveforms is employed.

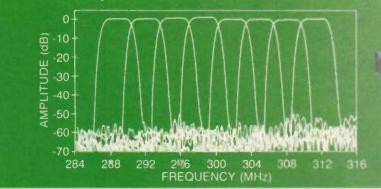
Designed for MIL-E-5400 and MIL-E-16400 environments, the limiter discriminator has an operating temperature range of -54C to +71C. It measures 3.53" × 1.50" × 0.47" excluding SMA connectors and weighs 3.7 oz. The specifica-



tions include a frequency accuracy of  $\pm$  12 MHz at 25°C and  $\pm$ 25 MHz above that temperature and a video sensitivity of 10 mV/MHz. The ICDX750 is priced at

\$1535. Other models are available with operating frequencies from 30 to 1000 MHz. RHG Electronics Laboratory, Inc., Deer Park, NY. INFO/CARD #219.





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#### **GaAs MMIC Amplifiers**

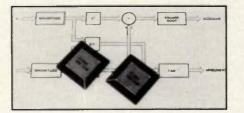
The UPG100 is a low noise GaAs MMIC and the UPG101 is a medium power GaAs MMIC. Both are internally matched to 50 ohms and operate in the .05 to 3 GHz range. The 100 has a gain of 16 dB with a maximum gain flatness of  $\pm$  1.5 dB, typical NF is 2.7 dB while the



power at 1 dB compression is 6 dBm. The specifications on the 101 include a gain of 14 dB, NF of 5 dB typical, and power at 1 dB compression is 18 dBm typical. The 101 maintains the same gain flatness as the 100. California Eastern Laboratories, Santa Clara, CA. Please circle INFO/CARD #218.

#### **Pythagoras Processor**

Plessey introduces a device that converts complex cartesian data into polar form magnitude and phase at a 10 MHz rate. Termed the pythagoras processor,



the PDSP16330 converts  $16 \times 16$  bit two's complement or sign magnitude data into 16 bit magnitude and 12 bit phase format. The magnitude may be scaled in amplitude by powers of two. Its applications include radar, sonar and digital radio systems. Plessey Semiconductors, Irvine, CA. INFO/CARD #217.

#### **Digitizing Scope With TDR**

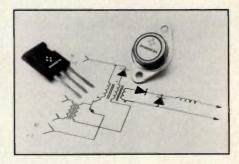
The HP 54120T color digitizing scope has a 20 GHz repetitive bandwidth, a built-in time domain reflectometer (TDR), and four simultaneous acquisition input channels. The features include optional 6 GHz passive probing, 12 bit resolution, .25 ps timing resolution and a 10 ps tim-



ing accuracy. The TDR measures impedance, reflection coefficient and distance from a reference plane. Hewlett-Packard Company, Palo Alto, CA. INFO/CARD #216.

#### **Power Switching Transistors**

General Semiconductors unveils a series of bipolar power switching transistors that offer a combination of 850 V ( $V_{CEV}$ ) and a switching speed of 35 ns (crossover time). They also have a  $V_{CEO}$  of 400 and 450 V and a maximum saturation  $V_{CE}$  of 1 V. These devices are designed to switch the inductive loads characteristic of switching power supplies.



characteristic of switching power supplies. General Semiconductor Industries, Inc., Tempe, AZ, INFO/CARD #215.

#### **High Speed Power Driver**

The UC3705 power driver uses a high speed Schottky process which allows it to interface between low level control functions and high power output devices. It is compatible with MOSFET and bipolar transistors. The device has a source/sink drive current of up to 1.5 A with a 40 ns rise and fall into 1000 pF. Unitrode, Merrimack, NH. INFO/CARD #214.

#### **Tunnel Diode Pulse Generator**

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### rf products Continued



amplitude, 4 us duration, and 50 kHz repetition rate. It consists of a pulse driver unit and a remote TD pulser head. The unit costs \$1950. Picosecond Pulse Labs, Inc., Boulder, CO. Please circle INFO/CARD #213.

#### 2 to 6 GHz MMIC Amplifier Chip

Harris Microwave introduces a dual stage, 2 to 6 GHz, 0.5 micron GaAs medium power MMIC in die form. It is cascadable and intended for gain stage applications. DC blocking on the RF output and bypass functions are incorporated. The maximum input VSWR is 2:1 while the typical noise figure at 120 mA is 6 dB. Harris Microwave Semiconductor, Milpitas, CA. INFO/CARD #212.

#### **GaAs Prescalar**

The MFG-8000 Series is a family of dual modulus (1/128, 1/129) high speed GaAs frequency dividers. The devices operate up to 1.6 GHz with a maximum power dissipation of 50 mW. The maximum output current is -5 mA. Mitsubishi Electronics, Sunnyvale, CA. INFO/CARD #211.

#### **Microwave Power Meter**

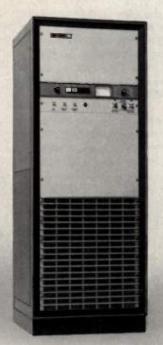
Wavetek introduces its 8531 single channel microwave power meter. It is complemented by a broad family of sen-



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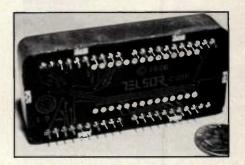
sors including thermocouple and diode sensors for high and low power measurements, low frequency sensors, 26.5 GHz sensors and 75 ohm sensors. The features include one button calibration, user controllable averaging of readings, and is programmable using GPIB format. Error is specified at  $\pm$  0.5 percent of full scale in the linear mode or  $\pm$  0.02 dB in the logarithmic mode. It can take up to 40 readings per second over the bus. Wavetek Corporation, San Diego, CA. INFO/CARD #210.

#### Portable Spectrum Analyzer

The Spectre 1050 spectrum analyzer offers portability through the use of self contained batteries that operate the instrument for three hours of continuous use. This 1 GHz unit has a line of options that include digital storage, tracking generator and a wideband video detector. Texcan Instruments, Indianapolis, IN. INFO/CARD #209.

#### **Interface Reader Module**

Telsor introduces an interface Model 2020 reader module. It is capable of multiplexing up to 8 addressable sensors which read 128-bit passive transponders



up to 12 inches away. The device measures  $2.6'' \times 1.2'' \times 0.7''$ . Telsor Corporation, Englewood, CO. Please circle INFO/CARD #208.

#### **L-Band Modular Synthesizer**

This OEM instrument is used to generate test frequencies of 1550 MHz to 1850 MHz in 1 MHz steps. Frequency control is provided by parallel TTL and CMOS input levels applied through a 25 pin D type connector. The specification include a phase noise of -85 dBc at 10 kHz, VSWR of 2:1, and a maximum settling time of 50 msecs. A sample of the reference frequency is available for coherent system operation. **Systematix, Lyndhurst, NJ. INFO/CARD #207.** 

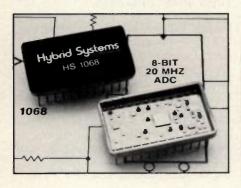


#### Quad OP Amp

NSC has released a quad operational amplifier that achieves a 140 dB dynamic range with less than 0.0015 percent distortion. The LM837 has a slew rate exceeding 8V/us, power bandwidth of 140 kHz and is able to drive 600 ohm loads. It has a low input noise voltage of 0.5 uV. It contains 4 amplifiers and is available in 14 pin dips and molded packages (SO). The price is \$1.25 each in 25,000 unit quantities. National Semiconductor Corporation, Santa Clara, CA. Please circle INFO/CARD #206.

#### 8-Bit A/D Converter

The HS1068 is a flash analog to digital converter that incorporates a wideband analog input amplifier, voltage reference and three stage output register into a 1.3"



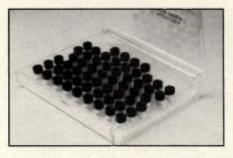
× 0.77" 24 pin dual in line package. It is an eight bit parallel flash converter capable of digitizing at rates up to 20 MHz. A digital output register provides three state outputs for ease of microprocessor interface. Hybrid Systems Corporation, Billerica, MA. Please circle INFO/CARD #205.

#### **GaAs Distributed Amplifier**

Texas Instruments introduces a monolithic GaAs amplifier that operates in the 2 to 18 GHz frequency range. The TGA8300 has a typical gain of 6.5 dB with an output power of 18 dBm at 1 dB gain compression. The noise figure is less than 7 dB over the frequency range. It includes on chip metal insulator metal bypass capacitors for the biasing of the amplifier. Texas Instruments, Inc., Semiconductor Group, Dallas, TX. INFO/CARD #204.

#### **Capacitor Evaluation Kits**

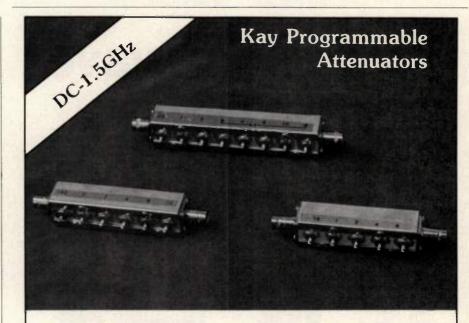
Monolithic ceramic chip capacitor evaluation kits are available from the Delcap Division of American Precision Industries. Each kit contains 25 samples in a range of values and are available in COG and X7R dielectric construction. Standard chip termination is palladium silver with



nickel barrier/solder available as an option. The rated voltage is 50 V. The kit costs \$129.95 for palladium silver and \$179.95 for nicker barrier. American Precision Industries, Inc., East Aurora, NY. INFO/CARD #203.

#### Synthesizer Receiver

Model SR500-F is a UHF synthesized FM receiver that operates at 420 to 470 MHz. The receiver is synthesized at 100



Kay Programmable Attenuators are offered in a variety of frequency and attenuation ranges for applications up to 1500MHz. Substantial discounts are offered on quantity (two or more units) orders.

Model No.	Imped- ance	Freq. Range	Atten. Range	Steps
P-1/4450	50	DC-1000MHz	0-16.5dB	.1dB
P-4460	50	DC-1500MHz	0- 31dB	1 dB
P-4480	50	DC-1500MHz	0- 63dB	1 dB
P.4450	50	DC-1500MHz	0- 127dB	1 dB
P.4440	50	DC-1500MHz	0-130dB	10dB
P-1/4457	75	DC- 750MHz	0-16.5dB	.1dB
P-4467	75	DC-1000MHz	0- 31dB	1 dB
P-4487	75	DC-1000MHz	0- 63dB	1 dB
P-4457	75	DC-1000MHz	0- 127dB	1 dB

Kay Elemetrics Corp manufactures a complete line of attenuators which includes Programmable, Standard In-Line, Miniature In-Line, Rotary (Bench and OEM) and Continuously Variable. For a complete catalog and price list or to place an order call Vernon Hixson at (201) 227-2000, Ext. 104.



Tel: (201) 227-2000, TWX: 710-734-4347 Kay Elemetrics Corp, 12 Maple Ave. Pine Brook, NJ 07058 USA



kHz channel spacings. Bandwidth is 360 kHz while the 12 dB SNR sensitivity is -95 dBm. The SR500-F is available with an analog or digital output. **General Ser**vices Engineering, Inc., Baltimore, MD. INFO/CARD #202.

#### Low Frequency Active Transponder

The Eureka 411 is a RF read-write iden-

tification device for use in data collection and identification applications. Up to 115 alpha numeric characters may be read from or written to the tag at a range of up to 39 inches from the interrogating antenna. The activating field is continuous wave at 132 kHz. Data from the tag is transmitted on a 66 kHz carrier using phase shift keying. Eureka Systems, Inc., Englewood Cliffs, NJ. Please circle INFO/CARD #201.

# **RF & Microwave Filters** 100 KHz to 15 GHz If you have a filter problem we'll provide the solution!



#### Hybrid VCXOs to 100 MHz

Reeves-Hoffman introduces a line of high frequency sine wave VCXOs in a double dip hybrid package. The frequency range is 2 to 100 MHz with power output to 10 dBm. Reeves-Hoffman Division, Carlisle, PA. INFO/CARD #200.

#### **30 MHz Measurement Receiver**

The Model VM-7 is a 30 MHz receiver designed for applications where precise measurements of RF power ratio are required over a wide dynamic range. It provides measurements over a range of 0 to 127 dBm, a selectable resolution of 0.1, 0.01 or 0.001 dB and an incremental attenuation accuracy of .02 dB per 10 dB. It is priced at \$15,000. Weinschel Engineering, Gaithersburg, MD. Please circle INFO/CARD #199.

#### **Power Silicon MMIC Amplifiers**

Avantek introduces two 100 MHz to 2 GHz Modamp™ cascadable gain blocks intended for use in narrow or broad bandwidth RF amplifier design. The MSA-0520 features +23 dBm power output at 1 dB gain compression and 8.5 dB gain at 1 GHz. The MSA-1023 features +27 dBm power output at 1 dB gain compression, +37 dBm third order intercept point and 8.5 dB gain at 1 GHz. Avantek, Inc., Santa Clara, CA. INFO/CARD #196.

#### Washable Trimming Capacitors

These miniature trimming capacitors feature internal "O" ring sealing which allow solvent washing. The TZ03 Series capacitance values are from 1.25 pF to 2.3 pF through 10 pF to 120 pF. The capacitors have an axle-less construction that contribute to its stable T.C. and small size. Murata Erie North America, Inc., Smyrna, GA. INFO/CARD #195.

#### **SMT** Capacitor Kit

NPO, X7R, and Z5U multilayer ceramic capacitor chips are furnished in Johanson Dielectrics' S-920 SMT prototype kits. It includes 550 capacitors of sizes 0805 and 1206 with nickel barrier terminations. It comes with a chip selector guide and a technical manual entitled *Understanding Chip Capacitors*. Johanson Dielectrics, Inc., Burbank, CA. Please circle INFO/CARD #194.

#### **PTFE Laminate**

Rogers Corporation introduces a line of woven glass reinforced PTFE laminates with dielectric constants ranging from 2.4 to 2.6 ± 0.04. Ultralam™ 2000 provides a highly uniform dielectric constant. Rogers Corporation, Chandler, AZ. INFO/CARD #193.

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# **rf** literature

#### **CMOS and D-MOS Data Book**

This data book from Topaz Semiconductor features data sheets and application tips on lateral N-channel enhancement and depletion mode D-MOS FETs, high voltage vertical N and Pchannel D-MOS FETs and high speed CMOS/D-MOS analog switches, multiplexers, and attenuators. It also provides a crossreference and substitution guide for identifying alternate sources, high performance plug in compatible components plus information on replacing JFETs with D-MOS FETs for significant increases in performance. **Topaz Semiconductor, San Jose, CA. INFO/CARD #189.** 

#### Software Literature

Compact Software has released literature on all its software products, including products formerly sold by Communications Consulting Corporation. Data sheets on Super-Compact revision 1.9 are available with details of all its additional capabilities. Data sheets on the company's line of microwave and RF, IBM PC, and Series 200/300 products called *CADEC 4* have been updated. **Compact Software, Paterson, NJ. INFO/CARD #188.** 

#### **Technical Note on Pulse Measurement**

Pulse Measurements in the Picosecond Domain is a technical note that discusses measurement problems, particularly those with emphasis in the picosecond domain. Oscilloscopes, calibration standards, coaxial cables and connectors, and probes are discussed. Also included is a list of references for extended reading on the subject. **Picosecond Pulse Labs, Inc., Boulder, CO. INFO/CARD #187.** 

#### **Application Notes on Mixers**

Synergy Microwave introduces a list of application notes and detailed information on surface mount technology for mixers. Synergy Microwave Corporation, Paterson, NJ. Please circle INFO/CARD #186.

#### **Microwave Components Fact Sheet**

Microwave Data Packet contains detailed fact sheets describing International Microwave Corporation's line of microwave components. Included are photographs, diagrams, descriptions and specifications for GaAs FET amplifiers, bipolar transistors and tunnel diode amplifiers, AGC and log video amplifiers, gunn diode and tunnel diode oscillators, broadband, narrow band and excess noise sources. International Microwave Corporation, Stamford, CT. INFO/CARD #184.

#### **Oscilloscope Support Material**

Tektronix had released a library of support material for its 2225 50 MHz oscilloscope. The material includes a demonstration video tape, brochure, a booklet entitled *The XYZ's of using a scope* and five technical briefs on oscilloscope measurements. The video tape demonstrates the 2225's dual channel operation, alternate horizontal magnification, 500 microvolt sensitivity and extensive triggering features. **Tektronix, Inc., Beaverton, OR. INFO/CARD #183**.

#### Surface Mounting Directory

IHS introduce a surface mount directory that enables the user to access over 30,000 devices which includes resistors,



#### **PRECISION CRYSTAL OSCILLATORS SERIES 8000 STANDARD FREQUENCY 5.0 MHz** MODEL ER8001 1 × 10 - 9/day AGING RATE MODEL ER8003 1 × 10 - 10/day MODEL ER8005 5 x 10 - 11/day PHASE NOISE SSB 1 Hz BW at 10 Hz offset MODEL ER8001 . . . 124 db MODEL ER8003 . . . 135 db INPUT VOLTAGE \_\_ 12 VDC ± 10% STANDARD OUTPUT \_\_\_\_\_ SINE-WAVE 1VRMS INTO 50 ohmLOAD SIZE MODEL ER8001 and MODEL ER8003 2" × 2" × 4" H MODEL ER8005 2.25" × 2.25" × 4.25" H MANY OPTIONS ARE AVAIL-**OPTIONS**

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# rf literature Continued

capacitors, inductors, transistors, bipolar and MOS digital/linear ICs, data arrays, diodes, and memory devices. The directory is available in 24X microfiche and 16 mm microfilm cartridges. It allows search by manufacturer's name, part number, generic part number or generic component description. The annual subscription rate is \$600. Information Handling Services, Englewood, CO. INFO/CARD #182.

#### Linear ICs Data Book

D.A.T.A. has released its 38th edition of *Linear ICs* which displays information for more than 15,000 linear ICs from 153 manufacturers. It has an expanded operational amplifier section with single, dual, triple and quad devices including programmable, ultra-low noise and BiFET types. Other devices in this book includes phase lock loop devices, amplifiers, sensor circuits, telecommunication devices, timers, transistor arrays, voltage comparators, and wideband devices. Two semiannual editions of this publication cost \$110. D.A.T.A., Inc., San Diego, CA. INFO/CARD #185.

#### **Linear Selection Guide**

Linear Technology Corporation introduces a brochure called Selection Guides. Among the bipolar and CMOS integrated circuits listed are single, dual and quad precision operational amplifiers, switching and positive fixed voltage regulators, low drop-out voltage regulators, positive adjustable and negative adjustable voltage regulators, voltage references, monolithic switched capacitor filters and various other components. Linear Technology Corporation, Milpitas, CA. INFO/CARD #180.



INFO/CARD 57

### **rf** software

#### Teledyne Microelectronics Offers Free Software

"RF Toolbox" is the title of a software package for everyday RF engineering calculations. Features include calculations for cascade noise figure and gain, resistive attenuators, inductive reactance and Q, capacitive reactance and D, resonance and tuning range, VSWR conversions, power level conversions and noise figure measurement. Also included are TSS calculations for DLVAs and TSS calculations with PD gain. The program is user-friendly and provides information on the equations used in the calculations. The software is free and users are entitled to free upgrades for one year provided they are registered with Teledyne. This promotional offer is intended to familiarize RF engineers with the custom hybrid development and manufacturing capabilities of Teledyne. Teledyne Microelectronics, Los Angeles, CA. INFO/CARD #174.

#### Synthesis and Analysis of Coaxial Transmission Lines

Microwave Software Applications introduces software that performs synthesis and analysis of coaxial transmission lines. Cxline performs synthesis of dimensions from impedance, analysis of impedance from dimensions, sensitivity analysis of impedance to incremental changes in inner conductor diameter, outer diameter conductor, and dielectric constant. The other functions performed include insertion loss versus frequency calculations for inner and outer conductor loss and dielectric loss effects. Higher-order mode cutoff frequency, capacitance and inductance per unit length, peak power handling of coax cross section and default values of each input parameter can also be calculated. Microwave Software Applications, Inc., Norcross, GA. INFO/CARD #173.

#### System Distortion Simulation

Designer of multichannel linear systems such as television repeaters, cable television distribution networks, microwave television links, and satellite transponders will find DPA-1000's distortion simulation useful in analyzing the interaction between signal to noise ratio and system linearity. DPA-1000 is ideal for frequency allocation studies as it can identify the source of intermodulation distortion in receivers. It calculates the second and third order distortion products arising from non linearity in multichannel systems. Version 1.1 of DPA-1000 read and writes data files compatible with BASIC, Fortran, and Pascal. It is priced at \$3500. Step Electronics, Inc., Campbell, CA. Please circle INFO/CARD #156.

#### Spice Plus-DEC VAX Interface

Analog Design Tools has integrated its Spice Plus™ program with the DEC VAX. After designing with the Analog Workbench, the simulation can be sent to a remote VAX for the processing. This allows the designer to use his workstation for the interactive design task, yet free up the workstation for other uses while the VAX is performing the simulation. The price of the remote Spice Plus starts at \$9,500 per MicroVAX. It will work in conjunction with Analog Workbench on Sun and Apollo workstations and with PC Workbench on IBM PC/AT. Analog Design Tools, Sunnyvale, CA. Please circle INFO/CARD #157.

# Antenna Design

Northrop Corporation's Defense Systems Division, located in sprawling Rolling Meadows, IL just northwest of Chicago, continues to provide innovation and leadership in its role as a major force in the electronic countermeasures industry. Our professionals contribute to the state-of-the-art within a creative, well-managed environment in which individuals are encouraged to develop their capabilities to the fullest, and work in concert as part of winning team. **Positions below require a BSEE, Physics or equivalent, MS preferred**.

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We seek a seasoned professional to spearhead our EW Antenna Design activities. Will be responsible for the design, development and transition into production of Airborne Antenna Systems for ECM applications, and the supervision of a growing unit of design engineers and technicians. Position will have frequent interface with manufacturing and program management. Requirements include a minimum of 12 years experience in airborne antenna design with at least 4 years of project level experience involving budgetary/scheduling responsibilities. Strong management and interpersonal skills are also essential.

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- Electromechanical Packaging
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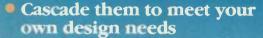
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Frequency (MHz)		Descri <mark>ption</mark>	Gain (dB) (Min.)	Noise Figure (d8) (Typ.)	Power Output at 1-dB Comp. (dBm) (Min.)	VSWR In/Out (Max.)	DC		
	Model <sup>1</sup>						Volts (Nom.)	mA (Typ.)	
50-500	6242-0501	LA7/LA7	22.0	9.5	+7.0	2:1	15	108	
	6243-0501	LA7/LA7/LA7	33.0	9.5	+7.0	2:1	15	162	
	6243-0505	LA7/LA7/LA7	33.0	9.5	+7.0	2:1	15	162	
10-1000	6242-0502	LA17/LA17	19.0	7.6	+5.0	2:1	15	110	
	6243-0502	LA17/LA17/LA17	28.5	7.6	+5.0	2:1	15	165	
	6243-0506	LA17/LA17/LA17	28.5	7.6	+5.0	2:1	15	165	
1000-4000	6242-0503	LA45/LA45-1	21.0	10.0	+14.5	2.3:1	15	210	
	6243-0503	LA45/LA45/LA45-1	30.0	10.0	+14.5	2.3:1	15	315	
2000-6000	6242-0504	KLA62/KLA62	20.0	9.5	+10.0	2.3:1	12	140	
	6243-0504	KLA62/KLA62/KLA62	30.0	9.5	+10.0	2.3:1	12	210	

Note: 1. These models may be cascaded together for wider limiting dynamic range.

For more information on these excellent products, contact Watkins-Johnson Company, Components Applications Engineering, at (415) 493-4141, ext. 2626.



Watkins-Johnson Company, 3333 Hillview Avenue, Palo Alto, California 94304-1204