

rf design

high frequency and analog engineering

September 1987



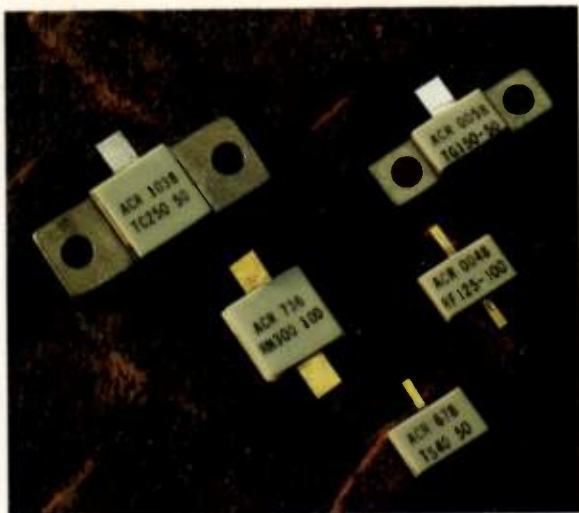
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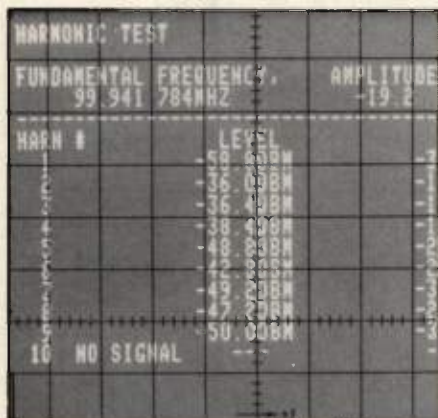
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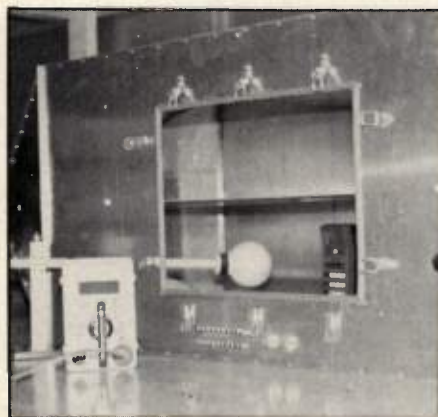
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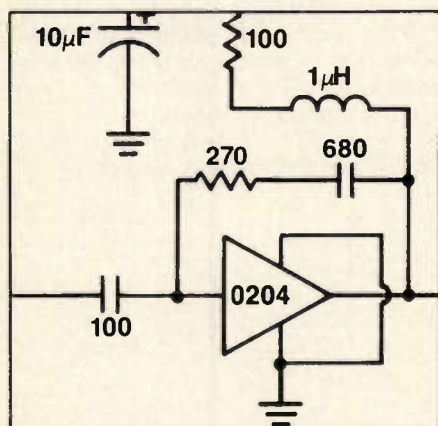
Small Signal Bipolars 



Page 39 — Programmable Instruments



Page 52 — EMC Test Labs



Page 62 — Low Cost Amplifier

Cover Story

27 Advances in Computer-Aided Test

New instrument capabilities, the low cost of personal computers, and advanced control and analysis software make automated RF testing more powerful than ever. EEsos's updated and improved ANACAT™ version 1.1 gets maximum flexibility and performance from automatic network analyzers. — *Larry Lerner*

Features

32 Special Report — Design Software for Personal Computers

Since personal computer hardware has become readily available to engineers, software has grown rapidly in both quantity and quality. This report contains a listing of RF and analog design software from 35 companies.

39 Featured Technology — Getting the Most Out of Programmable Instruments

Cheap and easy computing power has changed the way electronic instruments operate. The author fills us in on the history and present status of computer-controlled instruments and test systems for RF. — *Cliff Morgan*

52 RFI/EMC Corner — EMC Test Labs: Standards and Requirements

This article shows what an EMC test facility must have to perform both required and optional testing. Also covered is laboratory accreditation standards and qualifications of the staff. — *Mike Howard*

60 Designer's Notebook — PIN Diodes: Part III

Here is a design note reviewing the principles of transmit/receive (T/R) switching, following earlier lessons on shunt and series switch configurations. — *Andrzej Przedpelski*

62 A Low Cost 100 MHz-2.5 GHz Amplifier

This contest prize winner offers an inexpensive solution to wideband amplification with flat frequency response, using frequency-compensated silicon MMIC amplifiers. — *David Tharp*

65 A Noise Specification Program

This short BASIC program makes the conversions between various noise source parameters. — *James Mikeworth*

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rf editorial

The Computer — It's Only a Tool



Gary A. Breed
Editor

Not long ago on late-night TV, I saw one of those movies where a computer ran amok and took over the world. I laughed to myself about the unnecessary fears we once had about computers — taking jobs away from people, becoming smarter than we are, and finally getting rid of us. However, I have again grown uneasy about computers, now that PCs have cloned, networked and bussed their way onto most of our desks.

No, I'm not worried about computers taking over the world. I am concerned for the growing number of people who are allowing themselves to become computer-dependent. Regardless of the profession, many people are replacing their own ability with software. Sure, computers can store the experience and knowledge of the best human minds on any subject. Unfortunately, it is also possible for us to recover and use that information without ever understanding it.

I am worried about the student who got through college by writing term papers using word processing, spelling and style checkers instead of mastering the English language. Now he is an engineer who

plugs numbers into design programs without really understanding why they work. He certainly doesn't know which possible design solutions are better than others, he just takes the first one that comes up on the CRT.

What does he do when the power fails? Check your own staff next time the lights go out; see who heads for the door and who pulls out a calculator and proceeds with his work. (The ultimate problem is clearly dead calculator batteries. I have forgotten how to use a slide rule, and I refuse to look up logarithms in my old CRC Handbook!)

Don't get me wrong, computers are terrific! They are incredibly fast, and they're so dumb that they will do whatever we tell them to do. That's why we love them. They are our slaves, taking away the drudgery of repetition and providing valuable information when we want it. They let us do things in seconds that would take a week with pencil and paper. Consider doing an intermodulation analysis to the fifth order with multiple signal sources without a computer and you get the idea.

Computers allow us to spend more time being creative. The RF and analog design software listed in this month's Special Report is not intended to replace engineers with computers. These design aids have been developed to make good engineers even better, by letting them examine more design options, by modeling circuits before building prototypes, and by providing access to the expertise of the authors.

Most engineers understand these points. My concern is for those few who don't follow them — you know who you are. Don't get so dependent on computers that you can't function without them. Computers don't design, engineers do.

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INFO/CARD 5

M-1

'Megahertz Never Intended For Transmission'



By Keith Aldrich
Publisher

A couple of months ago, when writing about our newly-expanded charter (covering general analog as well as RF engineering — see July Viewpoint), I mentioned that we would reclassify our circulation to see how many of our *present* subscribers are *already* involved in "non-radio" applications of RF energy. In fact, that very issue had a revised subscriber card, to include such industries as computers and industrial controls for the very first time.

Well, about 18,000 of you have already returned the new card — enough so we can make some judgments about the rest of you as well — and sure enough! We have substantial numbers of readers in the "non-radio" areas. Around 2,000 of you work in computers and data transmission, for instance. That's more readers than we have working in land mobile radio

companies! About 1,000 of you are designing industrial controls and power supplies... more than those designing video and audio equipment.

These figures confirm what many of you wrote to us, in response to my editorial, all echoing more or less this comment from James Rieger, Naval Weapons Center, China Lake, Calif.: "Congratulations on acknowledging that there's Megahertz running around out there never intended for transmission!"

We also had three articles submitted — of a very high caliber, according to Editor Gary Breed — which their authors say they would never have sent to us if they hadn't read that editorial. The subjects of those three articles are (1) analog-to-digital converter evaluation, (2) active filter fundamentals, and (3) high-speed op amps. It looks as if you are in for a healthy strain of such topics in your *RF Design* diet.

Apparently there are plenty of you who need and will profit by it. But you don't need to be designing industrial controls for that to be true. Every phase-lock loop has a feedback control circuit, probably using an op amp. The majority of RF instruments today employ an analog-to-digital converter, in order to supply the digital features which make them so increasingly convenient and accurate.

The thrust of our editorial expansion is clearly a right move, even an overdue one. We look forward to a lot of excitement in carrying in out.

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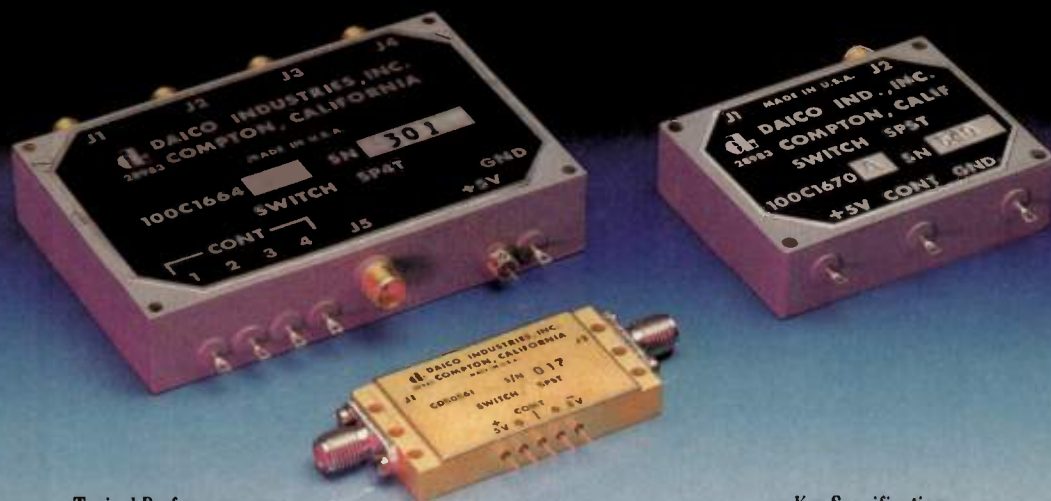
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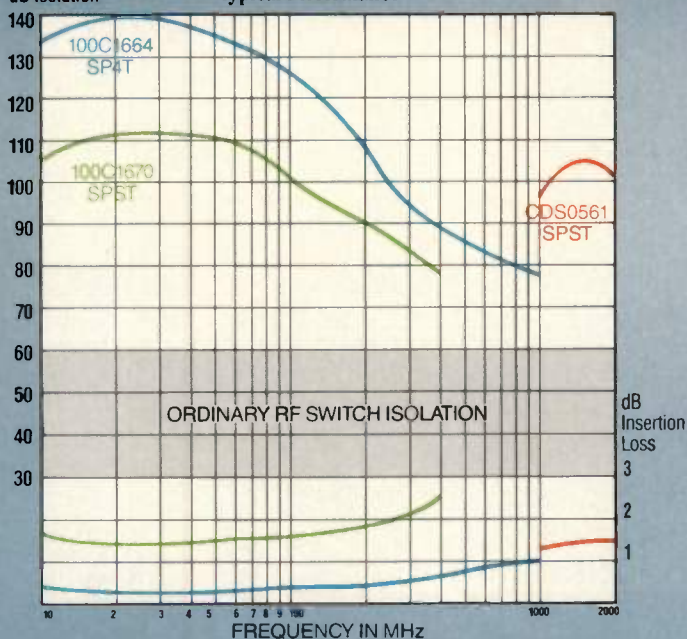
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Transition Time	4	2	—	ns	90%/10% or 10%/90% RF
Switching Speed	17	33	400	ns	50% TTL to 10%/90% RF
Insertion Loss	1.5	1.5	—	dB	—
	—	—	0.7	dB	20 - 350MHz
	—	—	1.0	dB	350 - 600MHz
Isolation	106	100	—	dB	—
	—	—	105	dB	20 - 200MHz
	—	—	85	dB	200 - 600MHz
VSWR	1.15 / 1	1.2 / 1	1.1 / 1	—	—
Termination VSWR	1.5 / 1	1.2 / 1	1.1 / 1	—	—
RF Power, Operate Max.	+13.5	+12	+24	dBm	0.1 dB Compression
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Operating Temperature	-55 - +125	-55 - +85	-55 - +85	°C	TA

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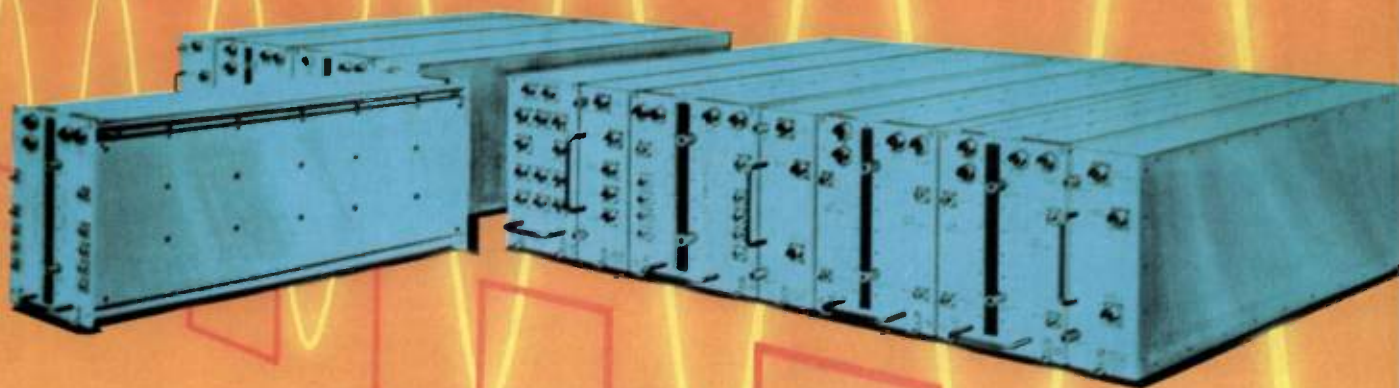
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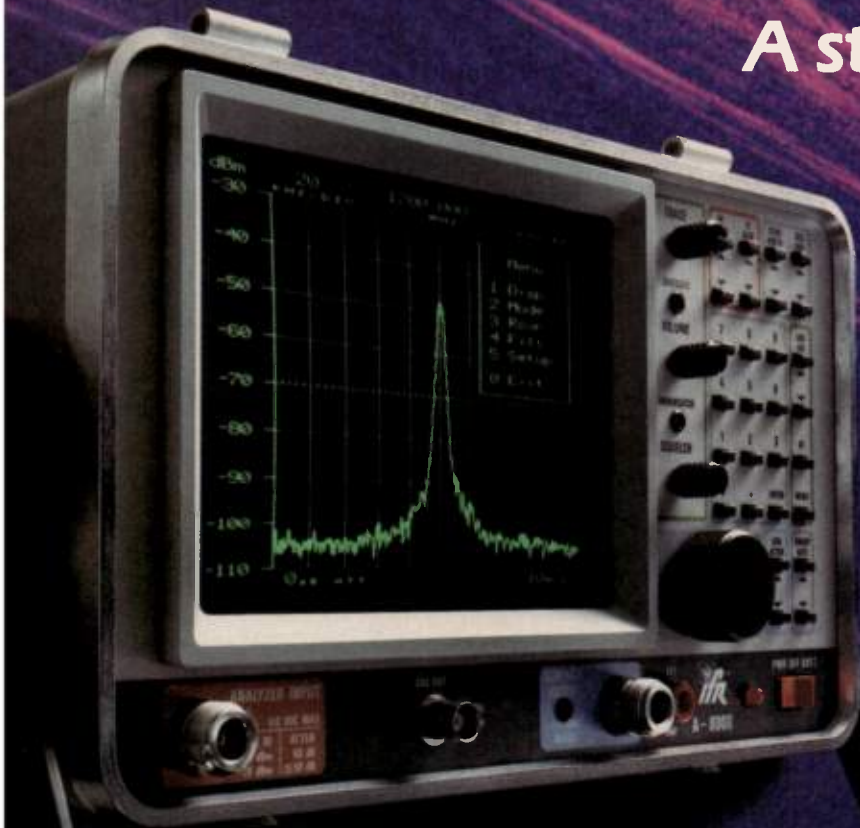
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INFO/CARD 8

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A Different Kind of Spectrum Analysis

Editor:

Remarkable fellow, that Isaak Newton (in the H-P ad). He got red light to refract farther than blue light!

James L. Engle
Phospho-Energetics, Inc.
Philadelphia, PA

We would like to point out this basic error to the artist whose work was used by Hewlett-Packard for their ad ("I'd settle for a bare bones spectrum analyzer. . ."). Unfortunately, he has been dead for almost 300 years. — Editor

A Glitch is Located

Editor:

There is an error in the program listing for "Computer Enhanced S-Parameter Design (February 1987, p. 105). In the sub-

routine for calculating C1, Line 410 is stated as:

410 I2 = -I(1)

whereas it should read:

410 I2 = -I(4)

The subroutine will now calculate $C1 = S11 - D^*S22$.

Joe Catalfamo
Somerset, NJ

There is one more error that readers should be aware of, on page 106. Line 810 should read:

810 IF K<1 THEN 1050

The "<" was omitted from the listing. — Editor

Another Update on Ladder Network Analysis

Editor:

I have received modifications to my Network Analysis Program (November 1986, *RF Design*) from readers Ronald Derencz and James Hanson, which may be of interest to those with early Commodore 64s, or IBM PCs with Hercules Graphics Cards.

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Some early C-64 computers used a different default color scheme which caused an apparent lack of graphics. In reality, the graphics plot was on the screen, but it was drawn using the same color as the background! In order to correct this, Mr. Derencz suggests the following changes:

- 1) Add the statement,
POKE P+54272,5 just before
NEXT P, in lines 2090, 2100, 2110
and 2120.
- 2) Line 2120: Delete the statement,
POKE 1224,79
- 3) Add a new line 2122 POKE
1224,79:POKE 54272+1224,5
- 4) Add a new line 2265 POKE
V+54272,7
- 5) Add a new line
2325 POKE V+54272,2
- 6) Save the modified program!

These changes will display a green border, yellow insertion loss and red return loss. The colors may be modified as desired by substituting different numbers for the 2, 5, and 7 above.

More and more IBM-PC clones are being sold with the Hercules Graphics Cards (or compatible). This presents a problem in that the network program does not plot with the correct scaling. Mr. Hanson provided the following information:

- 1) Add the following Hercules setup command within your AUTOEXEC.BAT file:
HGC FULL SAVE PRINT — You then need to re-boot your machine before running NETWORK4.BAS.
- 2) Start BASICA or GWBASIC and load NETWORK4.BAS.
- 3) Examine lines 2090, 2110, 2130, 2150, 2170, 2320, 2370, 2440, 2470, 2480, 2520, 2540, and 2550. Replace the existing scaling factors with new ones per the table below, as appropriate:

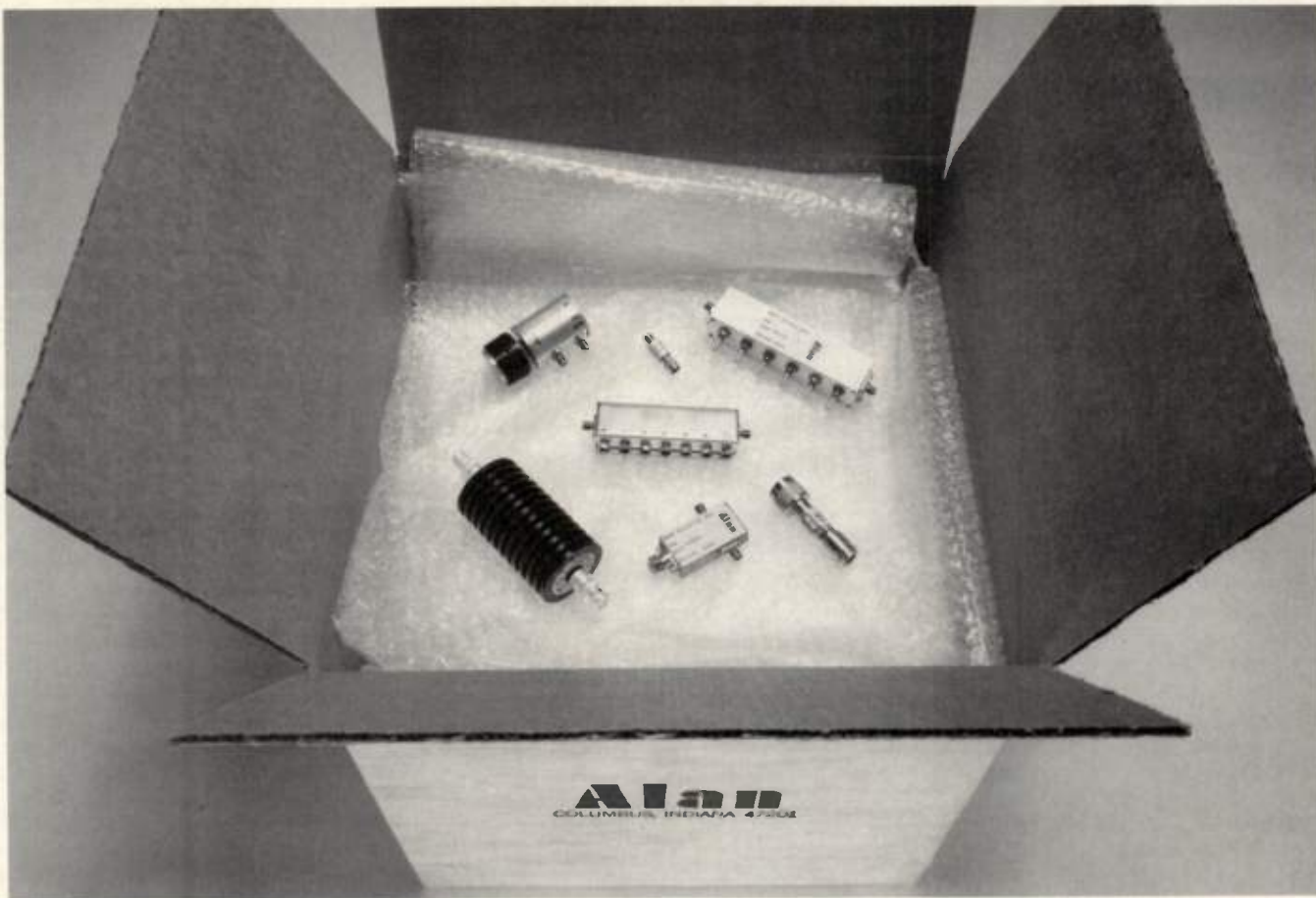
Old	New
20	35
30	52
100	113
140	244
150	261
160	278
500	563
600	675

- 4) Of course, save the modified program!

Since I do not own a Hercules card, I shall forward any requests for technical assistance to Mr. Hanson.

"Thank you for your support."

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Thomson, GE Make Business Swap

General Electric has agreed to sell its consumer products operation to Thomson S.A. for approximately \$800 million plus Thomson's assets in the medical electronic business. Most of the consumer products operation was acquired by GE in its merger with RCA. With this deal, GE will become a major force in medical electronics overtaking Siemens A.G. and Philips by obtaining GGR Medical Corporation of Baltimore, GGR Industries, GGR

MeV, and GGR Ultrasonics of France.

Thomson, on the other hand, will obtain a large manufacturing base for video and audio equipment which will be made up of plants in Bloomington, Ind., Indianapolis, Ind., and several Mexican assembly sites. Thomson's share of the deal comprises of GE's operations in 11 countries. Both companies will retain 19.9 percent in the businesses being exchanged. The deal is subject to government approval.

Magnavox Wins AVLS Contract

NAV-COM, Inc., a subsidiary of Magnavox Government and Industrial Electronics Company, has been awarded a contract by the town of Fairfield, Connecticut to install an Automatic Vehicle Location System (AVLS), which is to be used to track the movement and status of police cars and other emergency vehicles.

The system is expected to enhance the safety of field personnel, speed emergency response, ease dispatcher workload, enhance fleet control and management, reduce voice traffic on radio channels and improve procedures for reporting between

navigational satellites. At intervals, satellite data is used to recalibrate the vehicle's computed position, cancelling out errors that may have accumulated over time. The information from the processor is formatted and transmitted back to the dispatcher station via an existing radio channel where the location of each vehicle in the fleet is shown on a large map display.

Topaz Charts Its Course in DMOS

Double diffused MOS (DMOS) basically involves the diffusion of two types of impurities (N and P types) in areas where conventional circuits utilize only one or the other. It was invented at Signetics in 1968, but has been used in limited applications until Topaz Semiconductor was formed to focus exclusively on DMOS.

At a press conference on July 29, Thomas Cauge, president of Topaz, said that the market for DMOS will more than double over the next three years by expanding applications into low current digital and analog switching equipment. This includes DMOS-type devices now marketed by other companies under other names such as TMOSTTM (Motorola), HEXFETTM (International Rectifier) and SIPMOSTTM (Siemens). Cauge added that the total available market for all DMOS related devices will grow from \$380 million in 1986 to about \$443 million this year.

Topaz is now moving to more complex monolithic functions, using a two-layer metal process for wide applications in crosspoint video and RF switching, T-switches, and multiplexers. Also being developed are CMOS/DMOS thick film hybrid circuits which will be utilized in high frequency attenuators and adjustable filters.

Varian and IBM agree on Patent Licensing

Varian Associates, Inc., announced that it has reached an agreement with IBM Corporation which allows each company

use of the other's patents in their respective fields of interest. It is, however, limited to an exchange of patent licensing rights in the companies' current business areas. The agreement provides Varian with access to a broad range of patented technology, including developments in such areas as superconductivity and factory automation in device fabrication. Likewise, IBM will have access to Varian's patented semiconductor manufacturing technology for use in its internal chip manufacturing processes.

EEsof Users' Group to Meet at European Microwave Conference

The first European meeting of the EEsof Users' Group will be held in conjunction with the 17th European Microwave Conference at the Ergife Palace Hotel in Rome, Italy, September 7 through 11, 1987. The group will discuss applications, hear technical presentations, and exchange ideas about CAE software and microwave circuit design technology. Representatives from EEsof will be in attendance to answer questions and solicit suggestions for future product enhancement and developments. More information can be obtained by calling Michele Bush-Weller at EEsof, (818) 991-7530 or Helge Pecher, EEsof's European Center, West Germany, (49) 89-8577605.

Europe Sees EMC Boost

A report from Frost & Sullivan forecasts the European EMC market to expand from \$194 million in 1986 to \$227.5 million in 1987 and \$297 million in 1989. *Electromagnetic Compatibility Products and Services Market in Europe (#E892)* analyzes the market by materials, products and services, applications, OEM's and end-users, size, industry structure, and national characteristics. Many of the new opportunities taking shape as the market develops are



vehicles and dispatchers. The AVLS provides an automated means for tracking the location and status of moving vehicles at a central location. This is done by equipping each vehicle with a small navigation processor, which tracks its position automatically as the vehicle moves. The vehicles are fitted with a compact navigation processor which determine its geographic position by measuring the distance and direction traveled from a known starting point. This is accomplished with links from speed and heading sensors. The speed input comes from a transducer interface to the vehicles speedometer while the direction of travel is supplied by an electronic fluxgate compass. The vehicle's processor also includes a small receiver that uses signals from orbiting

Sales of Electromagnetic Compatibility Products & Services from Western Europe.

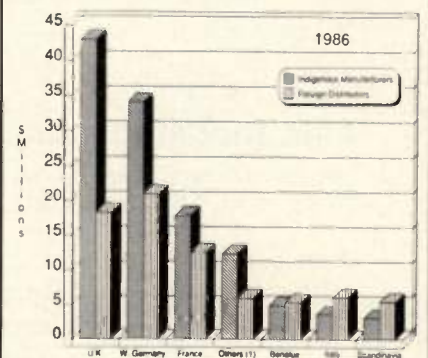
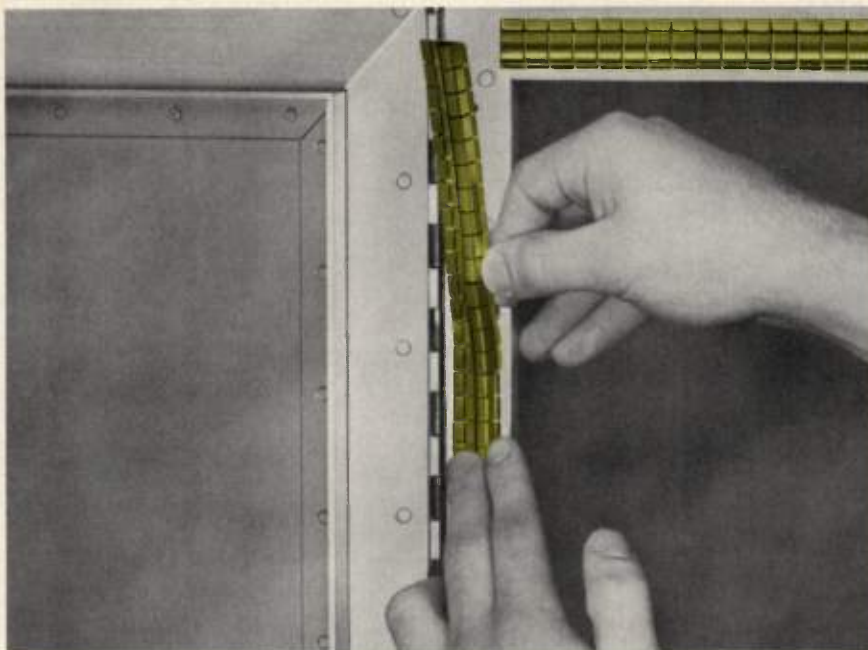


Figure courtesy of Frost & Sullivan.



New *stickn fingers*® design resists damage, provides superior RFI/EMI shielding!

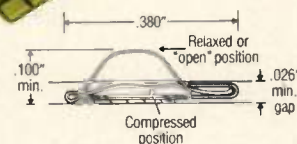
Now! Sticky-Fingers beryllium copper shielding gaskets have been made even better! A special, U-shaped end permits the upper part of the strip to slide when enclosure doors are closed, protecting against accidental damage to the strip's fingers.

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attenuation of more than 102 dB at 10 GHz plane wave,
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Complete information on these new strips, as well as the complete line of Sticky-Fingers shielding strips is available free. Write today to Dept. RF4.

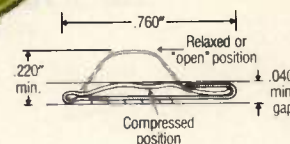
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Telex: 642-141 ANRITSU OKLD

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in design consultation, and testing and calibration.

Product categories which are predicted to have relatively high market growth rates include materials such as conductive coatings, and conductive polymers. The largest category in terms of sales, filters, and suppressors will maintain its lead going from \$39 million in 1986 to \$89 million in 1993. The report can be obtained from Frost & Sullivan, Inc. for \$2600. Call (212) 233-1080.

Analog Design Tools Joins Marketing Program

Analog Design Tools has joined EDA Systems' Synergy program along with nine other charter members. The Analog Workbench™ produced by Analog Design Tools will become part of EDA Systems' overall electronic design automation framework, called the Electronic Design Management System. Synergy is a cooperative marketing program for CAE products, created to provide customers with a wide selection of electronic design automation tools that take advantage of EDA's Electronic Design Management System Capabilities.

Coaxitube Corp. Produces Cable Assemblies

Coaxitube Corp. has commenced production of semi-rigid coaxial cable assemblies and delay lines at the firm's new Salisbury, Maryland facilities. This includes an RF testing laboratory with capabilities for measuring all types of flexible and semi-rigid cable assemblies in the DC to 26 GHz range.

Soladyne Enhances Testing Capability

By its recent acquisition of a Krautkramer Branson Ultrasonic Immersion Scanner, the Soladyne Division of Rogers Corporation has enhanced its capability to detect lamination flaws in bonded circuit boards. Using the instrument, Soladyne is able to test boards measuring up to 46" x 18" at both narrow band (for detection of very small, deep defects) and broadband (for good image resolution) frequencies.

Hytek Forms Military Product Division

Hytek Microsystems, Inc., has formed a Military Products Division aimed at diversification into the large market for hybrid circuits in defense and aerospace applications. The move follows the certification by the Department of Defense of Hytek's Carson City, Nevada facility to the requirements of MIL-STD-1772.

Eaton Awarded \$25M

Eaton Corporation has been awarded a \$25 million contract by the U.S. Customs Service for the development of a command, control, communications and intelligence (C³I) system to interdict drug smuggling along the southern border of the United States. The contract calls for the company to provide the government with the most current information-gathering and assessment technology available

in its fight against border drug trafficking.

Complementing the Eaton team is Harris Corporation, Government Electronic Systems Division, Melbourne, Florida, which is responsible for the system's broad range of communications technology and digital mapping capability. The system will be designed to receive real-time surveillance information from the nation's existing network of civilian and military air traffic control radars, as well

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as Customs Service marine radars and other sensing devices. Additionally, extensive intelligence data will be provided through other government agencies.

Keithley Completes Acquisition of Software Company

Keithley Instruments, Inc. has completed its agreement to purchase Adaptable Laboratory Software, Inc.. The action follows Keithley's recent purchase of Mac-

millan Software, a division of Macmillan, Inc., which marketed ASYST, ASYSTANT+, and ASYSTANT.

Hamilton/Avnet Announces New Facility

Hamilton/Avnet announces the opening of Hamilton/Avnet's new Cable Assembly Facility in Chatsworth, California. The facility stocks Amphenol/Spectra-Strip, Molex, AMP, TRW Cinch, and Winchester

Electronics products which are assembled to customers' specifications.

Hughes Receives \$300M

British Satellite Broadcasting has reached agreement with Hughes Aircraft Company on a \$300 million contract to provide the United Kingdom's first direct television broadcast by satellite. The service is expected to be in operation by late 1989. Hughes will build two satellites for BSB and deliver the spacecraft in orbit.

AEL Awarded Air Force Contract

AEL Industries, Inc. announced that its American Electronic Laboratories subsidiary has been awarded a \$5.6M contract by the U.S. Air Force to design, develop and produce UHF Satellite Terminal Antenna Subsystems (USTSAS) kits for C-130 and C-141 aircraft. The contract, awarded by the Warner Robins Air Logistics Center at Robins Air Force Base, GA, calls for AEL to produce 833 USTSAS kits with the first delivery scheduled for December 1987.

GE Division Gets Name Change

The commercial segments of GE's semiconductor business became GE Solid State effective August 3, 1987. The division, formerly known as GE/RCA Solid State, is responsible for the design, manufacturing and marketing of GE, RCA and Intersil semiconductors.

C for Programmers

Colorado State University announces a video tape course in C for programmers. The ten 30-minute video tapes are designed to enable experienced programmers to improve their skills in the high

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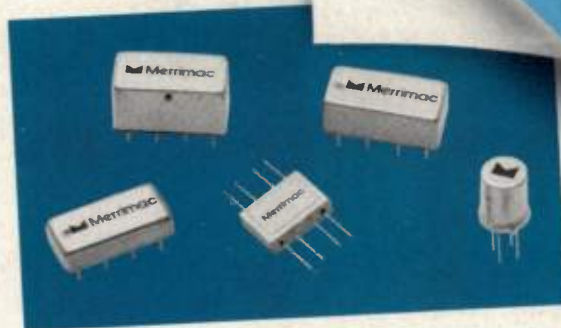
level C language. More information can be obtained from Pat Brown, Telecommunications Extended Studies, Division of Continuing Education, Colorado State University, Ft. Collins, CO 80523. Tel: (800) 525-4950.



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		RF/LO	IF				L-R	L-I	R-I				
/1-01S /1-01N	DMS-2-100/57855S DMS-2-100/57855N	0.05-200	DC-200	+4 min. +7 typ. +13 max.	6.5 typ. 8.5 max.	6.5	45 35	40 30	N/A N/A	0.05-30 30-200	+2	0	neg.
/1-02S /1-02N	117A/57834S 117A/57834N	5-500	DC-500	+4 min. +7 typ. +13 max.	7.0 typ. 9.0 max.	7.0	45 35	40 25	N/A N/A	5-50 50-500	+2	0	neg.
/1-03S /1-03N	DMS-2-250/57852S DMS-2-250/57852N	0.5-500	DC-500	+7 typ.	7.0 typ. 8.5 max.	7.0	35 30 25	30 25 20	N/A N/A N/A	0.5-1 1-250 250-500	+2	0	neg.
/1-04S /1-04N	DMS-8-500/57835S DMS-8-500/57835N	RF 2-400 LO 2-500	DC-800	+10 min. +20 typ. +23 max.	7.0 typ. 9.0 max.	7.5	40 35 25	40 35 25	N/A N/A N/A	2-32 32-100 100-500	+16	+14	pos.
/1-09S /1-09N	M119/57836S M119/57836N	1-750	DC-750	+7 typ.	7.5 typ. 8.5 max.	7.5	45 30 25	40 25 20	N/A N/A N/A	1-2 2-375 375-750	+2	0	pos.
/1-10S /1-10N	DMS-4-250/57853S DMS-4-250/57853N	0.4-500	DC-500	+7 min. +13 typ. +17 max.	7.0 typ. 9.0 max.	7.0	45 25	40 25	N/A N/A	0.4-50 50-500	+11	+8	neg.
/1-11S /1-11N	DMS-8-250/57837S DMS-8-250/57837N	1-500	DC-500	+23 typ.	7.5 typ. 9.5 max.	7.5	50 40 30 20	40 30 20 20	25 20 20 20	1-100 100-200 200-300 300-500	+16	+14	neg.
/1-12S /1-12N	DMS-2-25/57838S DMS-2-25/57838N	0.002-12	DC-12	+4 min. +7 typ. +13 max.	6.0 typ. 8.0 max.	6.0	45 40	40 30	N/A N/A	0.002-5 5-12	+2	0	neg.
/2-01S /2-01N	DMF-2A-505/57839S DMF-2A-505/57839N	5-1000	DC-1000	+10 min. +10 typ. +17 max.	7.0 typ. 8.0 max.	7.0	40 30	40 25	30 15	5-100 100-1000	+2	0	pos.
/2-02S /2-02N	DMF-2A-700/57840S DMF-2A-700/57840N	10-1500	DC-1000	+4 min. +7 typ. +13 max.	7.0 typ. 9.5 max.	7.0	30 25 25	30 20 18	N/A N/A N/A	10-600 600-1200 1200-1500	+2	0	neg.
/2-04S /2-04N	DMF-2A-250/57841S DMF-2A-250/57841N	0.5-500	DC-500	+7 min. +7 typ. +17 max.	7.0 typ. 8.0 max.	7.0	40 35	30 20	23 20	0.5-300 0.5-500	+2	0	pos.
/2-05S /2-05N	DMF-2A-250/57854S DMF-2A-250/57854N	0.5-500	DC-500 (IF-1 & IF-2)	+7 min. +7 typ. +13 max.	6.5 typ. 7.0 max.	6.5	35 30 25	30 25 20	25 15 15	0.5-10 10-200 200-500	+2	0	neg.
/7-01S /7-01N	M109/57832S M109/57832N	10-500	DC-500	+7 typ.	7.0 typ. 9.0 max.	7.0	40 35 30 25	35 30 25 15	25 20 15 10	10-50 50-100 100-200 200-500	+2	0	neg.
/7-02S /7-02N	M109/57833S M109/57833N	350-500	DC-500	+7 typ.	7.0 typ. 9.0 max.	7.0	25	15	10	350-500	+2	0	neg.

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Instrument Society of America

Fifth Measurement and Control Technology Short Course

October 5-9, 1987, Anaheim, CA

Information: ISA/87 Short Courses, Instrument Society of America, P.O. Box 12277, Research Triangle Park, NC 27709; Tel: (919) 832-5599

The George Washington University

HF Communications Technology

September 14-18, 1987, Orlando, FL

November 16-20, 1987, Washington, DC

Spread Spectrum Communication Systems

September 21-25, 1987, Washington, DC

Fiber Optics Technology for Communication

September 22-24, 1987, Washington, DC

Introduction to Modulators and Transmitters

October 5-6, 1987, Washington, DC

High Frequency Spectrum: New concepts and technologies

October 5-9, 1987, Washington, DC

Introduction to Receivers

October 19-20, 1987, Washington, DC

Modern Receiver Design

October 21-23, 1987, Washington, DC

Global Positioning System: Principles and Practice

November 4-6, 1987, Washington, DC

Wideband Communications Systems

December 7-11, 1987, Washington, DC

Information: Shirley Forlenzo, Continuing Education Program, George Washington University, Washington, DC 20052; Tel: (800) 424-9773, (202) 994-8530

UCLA Extension

Kalman Filtering II

October 26-30, 1987, Los Angeles, CA

Smart Armament/Missile Systems and Technology

October 27-29, 1987, Los Angeles, CA

Information: UCLA Extension, P.O. Box 24901, Los Angeles, CA 90024; Tel: (213) 825-1901; (213) 825-1047; (213) 825-3344

Interference Control Technologies, Inc.

Grounding and Shielding

September 15-18, 1987, Washington, DC

October 6-9, 1987, San Diego, CA

October 20-23, 1987, Boston, MA

November 3-6, 1987, Philadelphia, PA

Tempest Design

November 17-20, 1987, Mountain View, CA

Practical EMI Fixes

September 22-25, 1987, San Diego, CA

October 13-16, 1987, Washington, DC

November 10-13, 1987, San Jose, CA

Tempest Facilities

October 20-23, 1987, Washington, DC

MIL-STD-461/462

September 29-October 2, 1987, Washington, DC

RFI Design and Measurement

October 26-30, 1987, Philadelphia, PA

Information: Penny Caran, Registrar, Interference Control Technologies, Inc., State Route 625, P.O. Box D, Gainesville, VA 22056; Tel: (703) 347-0030

Integrated Computer Systems

Digital Signal Processing

September 22-25, 1987, San Diego, CA

October 27-30, 1987, Anaheim, CA

Image Processing and Machine Vision

September 22-25, 1987, Washington, DC

October 27-30, 1987, Palo Alto, CA

November 3-6, 1987, Toronto, Canada

Fiber Optic Communication

September 15-18, 1987, Washington, DC

September 29-October 2, 1987, San Diego, CA

November 3-6, 1987, Los Angeles, CA

November 3-6, 1987, Washington, DC

November 17-20, 1987, Boston, MA

Hands-On Programming in C

September 22-25, 1987, Washington, DC

September 29-October 2, 1987, San Diego, CA

October 27-30, 1987, Boston, MA

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Norand Corporation

MIL-STD 461B/C and 462 Regulations and Design Criteria

October 13-14, 1987, Cedar Rapids, IA

RF Susceptibility and ESD Testing

November 3-4, 1987, Cedar Rapids, IA

Information: Bev Reynoldson, Norand Corporation, 550 2nd St S.E., Cedar Rapids, IA 52401; Tel: (319) 846-2415

Compliance Engineering

Compliance Seminars: EMI, Safety, ESD, Telecom

September 29-October 2, 1987, Orlando, FL

October 27-30, 1987, Boston, MA

December 1-4, 1987, San Diego, CA

Information: Compliance Engineering, 593 Massachusetts Avenue, Boxborough, MA 01719. Tel: (617) 264-4208

Avantek, Inc.

RF & Microwave Semiconductor Design Solutions Seminar

October 2, 1987, Ft. Wayne, IN

October 5, 1987, Boston, MA

October 7, 1987, Ft. Washington, PA

October 9, 1987, Atlanta, GA

October 13, 1987, Orlando, FL

October 15, 1987, Dallas, TX

October 19, 1987, Burbank, CA

November 4, 1987, Albuquerque, NM

Information: Janice Little, Avantek, Inc., M/S 3G, 3175 Bowers Ave., Santa Clara, CA 95054. Tel: (408) 970-2139

R & B Enterprises

Electromagnetic Pulse (EMP) Design and Test

October 5-6, 1987, Philadelphia, PA

November 17-18, 1987, Washington, DC

December 7-8, 1987, Philadelphia, PA

Electromagnetic Pulse Workshop

October 9, 1987, Philadelphia, PA

Understanding and Applying MIL-STD-461C

October 20-21, 1987, Philadelphia, PA

November 17-18, 1987, Washington, DC

December 14-15, 1987, Philadelphia, PA

MIL-STD-461C Praxis (Workshop)

October 20-21, 1987, Philadelphia, PA

Information: Greg Gore, Director of Training, R & B Enterprises, 20 Clipper Road, West Conshohocken, PA 19428. Tel: (215) 825-1960

September 29-October 1, 1987

1987 OEM Design and Integrated Manufacturing Exposition
McCormick Place East, Chicago, IL
Information: Bill Little, Penton Exposition, 122 East 42nd St., New York, NY 10168; Tel: (800) 634-4639, in NY call (212) 867-9191

September 28-October 1, 1987

Ninth Annual Meeting and Symposium, Antenna Measurement Techniques Association
The Stouffer Madison Hotel, Seattle, WA
Information: James R. Otey, 1987 AMTA Symposium, 6632 South 191st Place, Suite E-105, Kent, WA 98302

September 29-October 1, 1987

Fall National Design Engineering Show and Conference
Javits Convention Center, New York, NY
Information: Monica Viladegutt, Cahners Exposition Group, 999 Summer St., P.O. Box 3833, Stamford, CT 06905; Tel: (203) 964-0000

October 4-8, 1987

ISA/87
Anaheim Convention Center, CA
Information: Paul Albert, Instrument Society of America, P.O. Box 12277, Research Triangle Park, NC 27709

October 12-14, 1987

GaAs MAN-TECH Conference
Marriott Hotel, Portland, OR
Information: Wayne Moyers, Pacific Monolithics, 245 Santa Ana Ct., Sunnyvale, CA 94086; Tel: (408) 732-8000

October 13-15, 1987

Scan-Tech 87
Bartle Hall, Kansas City, MO
Information: Don Anderson, AIM, Inc., 1326 Freeport Rd., Pittsburgh, PA 15238. Tel: (412) 963-8588

October 19-21, 1987

The Twentieth Annual Connectors and Interconnection Technology Symposium
Franklin Plaza Hotel, Philadelphia, PA
Information: Electronic Connector Study Group, Inc., P.O. Box 167, Fort Washington, PA 19034-0167; Tel: (215) 825-3840

October 26-30, 1987

FOC/LAN 87, Eleventh International Fiber Optic Communication and Local Area Networks Exposition
Anaheim Convention Center, Anaheim, CA
Information: Renee Farrington, Information Gatekeepers, Inc., 214 Harvard Avenue, Boston, MA 02134; Tel: (617) 232-3111

November 10-12, 1987

1987 ITEA Symposium
Park Plaza Hotel, Boston, MA
Information: Howard L. Graves, Raytheon Company, Public Relations, 141 Spring Street, Lexington, MA 02173; Tel: (617) 470-6027

November 11-13, 1987

RF Expo East 87
World Trade Center, Boston, MA
Information: Linda Fortunato, Convention Manager, Cardiff Publishing Company, 6300 S. Syracuse Way, Suite 650, Englewood, CO 80110; Tel: (303) 220-0600; (800) 525-9154

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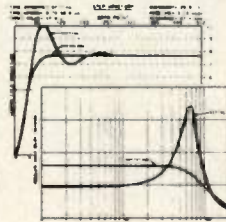
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Advances in Computer-Aided Test

PCs and Advanced Software Maximize Instrument Capabilities

By Larry Lerner
EEsof, Inc.

Today's test and measurement equipment continues to advance the state of the art in reliability and measurement capabilities, evidenced by Hewlett-Packard's 8510 and 8753 network analyzers, plus the newest entry on the market, the Wiltron 360™ vector network analyzer. These test instruments (pictured on the front cover) address measurements in the RF/microwave range for both active and passive networks over the 100 kHz to 26.5 GHz frequency range, and in the case of the Wiltron 360, up to 40 GHz. Recent computer-aided test (CAT) software such as ANACAT™ 1.1 from EEsof, Inc., has been created to give the RF engineer maximum performance in the calibration, measurement, and management capabilities of this test equipment in the lab.

As little as ten years ago, computer-aided test was confined to the lab where the engineer made measurements with custom-made software programs running on desktop instrument controllers. The process was cumbersome and the results often incomplete. Data had to be manually transferred to simulator programs running on large mainframe computers. The data transfer from test equipment to instrument controllers to program-mable software was slow and inexact.

PCs Bridge the Gap

With the advent of the IBM PC™ in 1981 and the subsequent development of software customized specifically for this environment, the requirements for gathering, processing, and measuring test data for use in future simulations were bridged. The PC offered a powerful, portable, and versatile computing environment that could be utilized right in the lab, replacing instrument desktop controllers and the VAX mainframes that provided the early simulations, and lowering the price of an average workstation from the 50-60K range to an affordable 7-10K. As with every other industry, the PC had a revolutionary impact on RF/microwave circuit design and test — from an increase in produc-

tivity and creativity for the engineer to huge time and cost reductions in production.

ANACAT 1.1

ANACAT was written from the ground up to maximize the features of the workstation and test equipment pictured below, and the new, upgraded ANACAT 1.1 release has additionally been ported to the 32-bit HP 9000 series 300 computers. It not only commands the full capabilities of the HP test equipment but a full family of network analyzers, including the advanced capabilities of the Wiltron network analyzer. ANACAT provides a flexible and interactive means of controlling the operation of these network analyzers, and for acquiring, correcting, manipulating, displaying, or managing the collected data. This interface provides engineers with a tool to facilitate in-depth measurement of both the pre-design and

production test phases of circuit engineering.

The PC version of the program has a DOS interface, full-screen editor for creating and revising files, and color graphics for interactive graphics display. In addition, the program's database readily interfaces with other popular applications such as EEsof's Touchstone® program for circuit simulation, Lotus' 1-2-3™, and Ashton-Tate's dBASE™. New capabilities created for version 1.1 include a new time-domain output feature, macro commands, enhanced view menu, and multi-port measurements.

Measurements and Database Manager

The heart of the program is the measurement database, which is characteristically a task beyond the scope of the automatic network analyzers' firmware. ANACAT takes large amounts of data



ANACAT 1.1 extends the calibration capabilities of automatic network analyzers.

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117CFR	100 W	100-160MHz	45dB	100-240V	35dB	3,495.00
155LCR	100 W	6KHz-10MHz	48dB	100-240V	25dB	3,250.00
161C	100 W	.5-90MHz	48dB	100-240V	23dB	3,995.00
162LP	100 W	50-200MHz	50dB	100-240V	23dB	4,995.00
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118FC	125 W	225-400MHz	45dB	100-240V	22dB	4,880.00
121CA	200 W	.5-35MHz	50dB	100-240V	22dB	3,250.00
121CR	200 W	2-30MHz	50dB	100-240V	50dB	3,690.00
122C	200 W	10KHz-220MHz	50dB	208-380V	17dB	9,550.00
123CFR	200 W	80-160MHz	50dB	100-240V	35dB	5,850.00
126CR	200 W	225-400MHz	50dB	100-240V	22dB	7,500.00
127CR	200 W	400-500MHz	50dB	100-240V	50dB	8,500.00
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164SP	250 W	400-500MHz	55dB	100-240V	25dB	6,990.00
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130CF	400 W	80-160MHz	50dB	100-240V	30dB	8,990.00
131CF	400 W	225-400MHz	50dB	100-240V	23dB	18,500.00
128CF	500 W	2-30MHz	55dB	100-240V	16dB	7,250.00
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134C	500 W	10KHz-220MHz	60dB	208-380V	17dB	17,990.00
134CM	500 W	1-200MHz	60dB	208-380V	17dB	17,100.00
164LP	500 W	10-86MHz	53dB	100-240V	23dB	6,990.00
164HP	500 W	80-180MHz	53dB	100-240V	23dB	7,990.00
166UP	600 W	200-400MHz	55dB	100-240V	25dB	8,990.00
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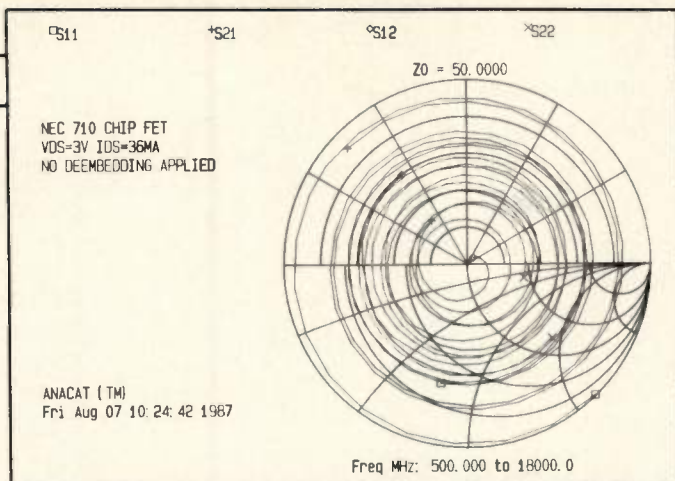


Figure 1. A full, 2-port calibrated measurement of a NEC 710 FET including the test fixture. No de-embedding has been applied by ANACAT at this point to remove the effects of the test fixture.

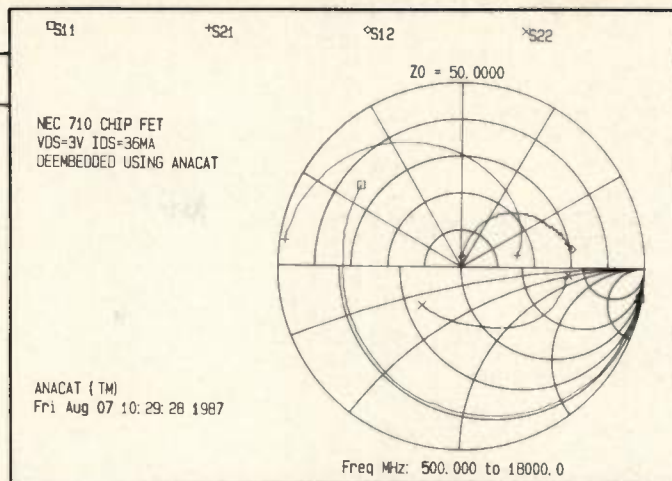


Figure 2. This shows the same NEC 710 FET measurement data from Figure 1, with the test fixture effects removed using the ANACAT de-embedding feature. All four de-embedded FET S-parameters are graphically displayed on a combination Smith-Polar chart.

directly from the network analyzer and stores and manipulates the data in a structured yet flexible database. It's a hierarchical database in which the user can organize measurement data with customized labels, descriptions, and so forth; and the database manager can be readily customized in terms of how the user can organize measurement data.

The database manager lets the user create, delete, and revise test data with a few simple keystrokes. The Lotus/dBASE

compatibility opens the door to more direct information exchange. Future product updates will offer a Measurement Data Interchange Format (MDIF) that adds consistency in viewing and flexibility in manipulating the large amounts of stored data in the system.

The database allows the data to be saved and viewed at a later time in such forms as S-, Z-, Y-, G-, and H-parameters. The data is then displayed on a customized selected table or graph in the desired

format.

Other features of the program's database are its ability to normalize data in correspondence to any other given measurement and to view calculated parameters relating to the measurement, such as maximum available gain and stability. Using Touchstone, the user can access measurements from the database and apply them either in a simulation or to fit the measured data to equivalent circuit models. Rectangular and polar charts

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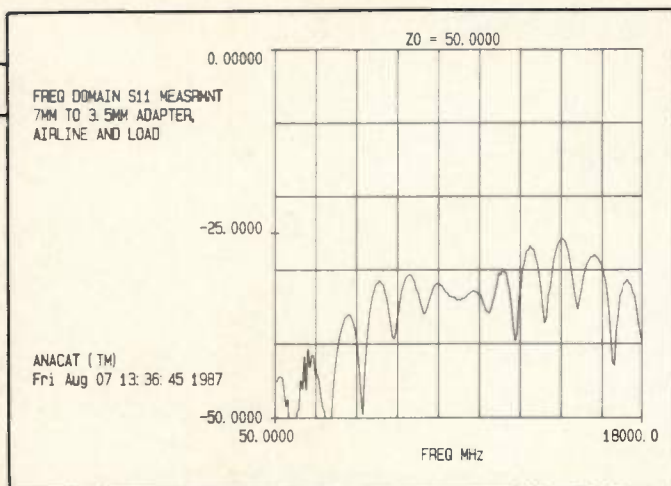


Figure 3. Frequency-domain S11 response of a 7 mm to 3.5 mm adapter, airline, and load combination from 50 MHz to 18 GHz. In the frequency domain, it is not possible to separate the mismatch effects of the 35 mm adapter and airline from the load termination.

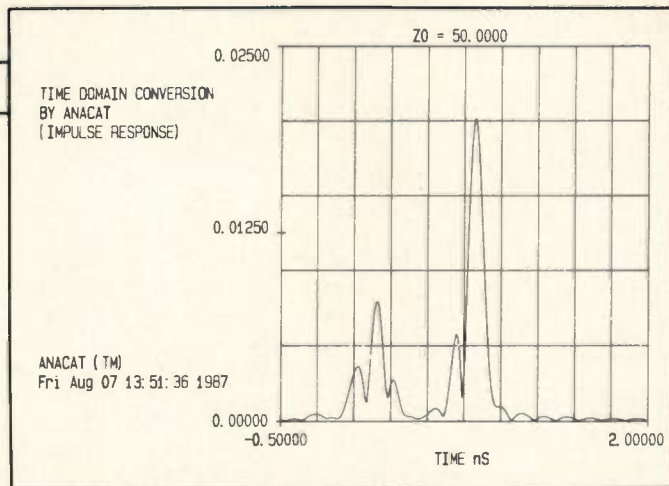


Figure 4. The frequency-domain response is converted to the time domain using ANACAT. Now the effects of the 3.5 mm adapter and airline are readily apparent and easily measured. Position instructions... Figs. 1 and 2 need to be adjacent, as do Figs. 3 and 4.

may be customized for comparison of data dispersed throughout the database.

The program will completely define a calibration of the network analyzer system. This is accomplished in an interpretive English language and is easy to understand and modify, providing calibration standards definition using Touchstone-like syntax. Calibration standards are defined using models of coaxial, waveguide, or planar transmission media, and terminating impedances may be ex-

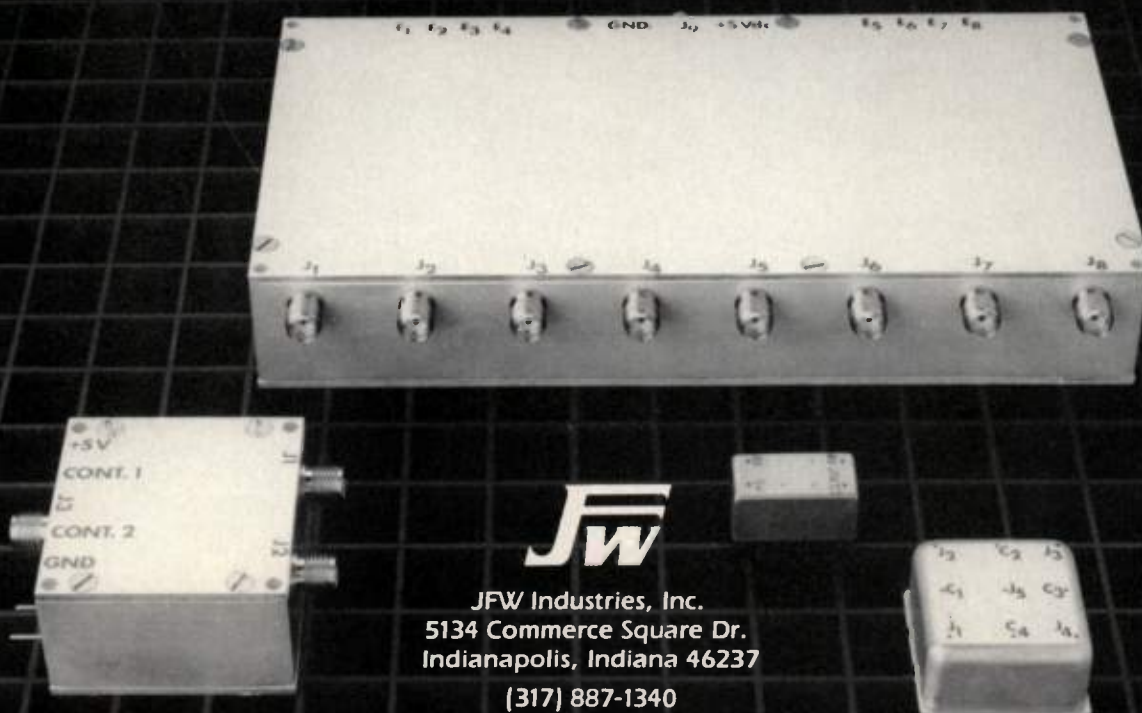
pressed in general equation formats.

Using the interactive menus, the engineer can measure customized standards and derive associated network analyzer error coefficients. The definition file standards measurements and derived error terms can be stored and retrieved as needed from the calibration database. Unlike manual operation of the vector network analyzer, ANACAT defines and implements a calibration at any arbitrary collection of up to 512 frequencies.

Embedding and De-Embedding Data

In numerous measurement applications such as for active devices embedded in a planar transmission media, it is difficult to manufacture standards which can, with confidence, be modeled for use in standard error correction. ANACAT enables the engineer to calibrate the network analyzer at a convenient set of reference planes and to then math-

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ematically rotate or transform those planes through network descriptions provided by the user. These network descriptions can originate from the program's measurement database, from standard Touchstone S-parameter data files, or can be extracted directly from measurement of unit fixture standards.

For example, if the engineer takes a calibrated measurement from the network analyzer and would like to see how that unit acts when embedded in a hypothetical circuit, ANACAT rotates the calibrated reference planes out or away from the device through the hypothetical networks. Successive application of de-embedding and then embedding operations allows real-time measurement of active devices in hypothetical circuits.

In the case of simulators, most work is done in the frequency domain. But if an engineer wants to understand a circuit physically, then at each juncture of the connected circuit elements it's often advantageous to analyze the circuit in the time domain. Underlying the behavior of the S-parameters with frequency are the cumulative effects of a series of reflected signal waves created at each discontinuity

within the circuit. ANACAT allows observation in the time domain of these physical effects in terms of the distance between discontinuities in the circuit, and it lets the user de-embed based on that time-domain information.

ANACAT lets the engineer view any display function as a function of time and to observe time-domain output in either RF or baseband format, for either impulse or step response. The user can de-embed a set of reference planes in the frequency domain by internally converting them to time domain, gating the time domain, and then re-converting to the frequency domain.

The current network analyzers can measure one or two port devices, and that's a limitation. To measure something that has more than two connections necessitates alternate measurements by connecting two ports at a time.

One can mathematically supply three-four- or five-port S-parameters based upon composite measurements for the different two-port connections. ANACAT provides a shorthand way that lets the engineer measure and view measurements of up to six-port devices.

Conclusion

Today's RF engineer has workstations available to him or her that have, in the past few years, dramatically fallen in cost and risen in performance. Information from powerful network analyzers can be easily acquired and manipulated in program databases such as ANACAT for all phases of design and production.

In the future, we will witness the incorporation of artificial intelligence in CAT to produce much more sophisticated and flexible databases and operating systems. The future for RF and microwave design offers unlimited possibilities.

For more information about ANACAT 1.1, contact EESof, Inc., 31194 La Brea Drive, Westlake Village, CA 91362, tel. (818) 991-7530, or circle INFO/CARD #125.

About the Author

Larry Lerner is manager of CAT development at EESof. He is currently developing interactive test software and is the designer of ANACAT. He holds a BSEE from Harvey Mudd College and an MSEE from Stanford University.

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A Complete Listing of RF and Analog Engineering Programs

Radio frequency circuits and related areas of analog design may be the last engineering disciplines to receive widespread computer software support. It isn't that the field has been neglected. Rather, it is the difficulty of creating accurate mathematical models that slowed development. RF design has traditionally been a matter of approximation, refined to a final form through experimentation. The image of RF engineers as practitioners of a "black art" is still strong.


As you can easily tell by the amount of software now available, there is no longer any shortage of computer assistance for the RF engineer! The listing included in this report contains 35 companies and 108 different program packages. There is more available than just this listing, too. A few companies may

have been omitted (please tell us about them), and there is plenty of non-commercial software developed by individual engineers and distributed via bulletin boards, magazine articles, or passed along person-to-person.

The wait for RF software has been worth it. Many of the old problems of defining component characteristics, establishing correct transfer function equations and modeling stray reactance have been solved, or at least closely approximated. Designers can now achieve good correlation between software circuit analysis and "real world" performance. This is particularly true in offerings from the eight or ten medium to large sized companies who have invested significant time and money for development.

There are also many small companies who offer low cost software to engineers.

Although these programs may not be as sophisticated or have graphics as nice as those from bigger firms, they represent a good dollar value. These programs are usually solutions to a specific engineering problem, and often get used the same way as a reference book. Many engineers find excellent value in having such a reference collection on disk.

Whether from large companies or small, the amount of RF software is growing rapidly, especially programs for personal computers. Now that software designers are achieving good RF circuit models, engineers are gaining confidence in software tools that enhance their design capabilities. The following list (in random order) represents all the RF and analog design software for personal computers that the *RF Design* staff has been able to identify. 

PC-based RF and Analog Design Software

Analog Design Tools

The *PC Workbench™* simulates the design process from drawing schematics through simulated breadboarding, testing, and performance analysis.

Simukit™ allows the integration of the Workbench with users' simulators and model libraries. **Analog Design Tools, Inc., Sunnyvale, CA. INFO/CARD #160.**

Sofcad Electronics

Linear Circuit Analysis and Design, *LINCAD*, performs design, optimization, frequency and phase graphics, and various other functions.

CALCAD is a logic circuit analysis and design program.

FFTSA and *CONVOLV* are spectrum and signal analysis programs. **Sofcad Electronics, Inc., Columbus, OH. Please circle INFO/CARD #159.**

Thomas H. Stanford Engineering

RFAMP designs RF single stage amplifiers using S-parameters.

MICRO designs asymmetric couplers, end coupled band-pass filters and various other systems for microstrip design.

Thomas H. Stanford Engineering, Escondido, CA. Please circle INFO/CARD #158.

Microwave Software

Sepectre™ 2.0 is a frequency domain circuit analysis program.

OptiMatch™ 2.0 performs circuit optimization.

SmithMatch™ 2.0 allows the user to analyze or synthesize impedance match networks.

Utilities+™ is a collection of 25 RF/microwave circuit design tools.

MStrip+™ 2.0 is a microwave transmission line synthesis

and analysis program. **Microwave Software, San Juan Capistrano, CA. INFO/CARD #157.**

DGS Associates

S/FILSYN™ is structured for the design, synthesis, and analysis of passive LC, active RC, digital and microwave filters, and delay equalizers.

COMBLINE is a filter post processor program that works on a design prepared by *S/FILSYN*.

SCASY is a program for analysis of linear switched capacitor circuits.

Filter-9 PAC contains nine programs for the design of passive LC filters and transfer of active RC and digital filters. **DGS Associates, Santa Clara, CA. INFO/CARD #156.**

Interceptor Electronics

ALADYN-64 is a circuit analysis program for examining the characteristics of linear and ladder networks. **Interceptor Electronics, Inc., Round Hill, VA. INFO/CARD #155.**

Webb Laboratories

Transcad is a transmission line and waveguide synthesis/analysis package.

Syscad performs link analysis, architecture characterization, optimization, receive and transmit spurious analysis, and frequency source phase noise analysis.

Filsolve performs synthesis on filters in Butterworth, Chebyshev, Bessel, and equal ripple time delay realizations. **Webb Laboratories, North Lake, WI. INFO/CARD #154.**

Multipath

Public Domain & User Supported Software Directory lists approximately 3200 programs which include science, engineering, and math. **Multipath, Inc., Montville, NJ. Please circle INFO/CARD #149.**

RLM Research

Active Filter Design synthesizes and analyzes active filters.

Spice File Conversion Utility Program simulates filters designed with *Active Filter Design* using Spice without having to manually create the input file. **RLM Research, Boulder, CO. INFO/CARD #153.**

Etron RF Enterprises

RF Notes 1 performs general design aids which include image parameter filters, basic stripline design, and basic microstrip design.

RF Notes 2 performs attenuator pad design, impedance matching, inductor design, and capacitor analysis.

RF Notes 3 Vol 1 designs Butterworth filters.

RF Notes 3 Vol 2 is tailored for Bessel filter design.

RF Notes 3 Vol 3 performs Chebyshev filter design.

RF Notes 4 is a network analysis program that creates circuit design with up to 30 circuit sections. **Etron RF Enterprises, Diamond Bar, CA. INFO/CARD #152.**

Hayward Electronic Systems

Ladpac/Ladpac-87 features filter design, circuit analysis, optimization, error trapping, impedance matching, and various other functions. **Hayward Electronic Systems, Inc., Beaverton, OR. INFO/CARD #151.**

BV Engineering

ACNAP is a circuit analysis program.

COMCALC is a communications system design package.

LCFIL performs passive lumped RF filter design.

PCPLOT provides high resolution screen/printer graphics.

BV Engineering Professional Software, Riverside, CA. INFO/CARD #150.

Circuit Busters

Filter synthesizes thirteen topographies of lowpass, high-pass, bandpass, elliptic lowpass, and elliptic bandpass filters.

Superstar 3.0 is a general purpose RF and microwave circuit simulation and design tool. **Circuit Busters, Inc., Stone Mountain, GA. INFO/CARD #148.**

Intusoft

Pre_Spice is an interactive circuit analysis program that performs Monte Carlo analysis among various other functions.

Is_Spice performs transfer functions and noise analysis, transfer function sensitivity and curve families, and various other functions.

Intu_Scope is a graphics processor that displays data obtained from *Is_Spice*. **Intusoft, Rancho Palos Verdes, CA. INFO/CARD #147.**

Microwave Software Applications

SLINE performs synthesis and analysis of single-strip stripline.

MLINE is a software package for the synthesis and analysis of single-strip microstrip transmission lines.

CXLINE is tailored for coaxial transmission lines.

LANGE offers microstrip interdigital coupled software.

SLCUP designs edge-coupled stripline directional couplers.

MSCUP performs design synthesis and analysis of edge coupled microstrip directional couplers.

LEFLTR is a lumped element filter program.

CXLPF designs distributed circuit lowpass filters in coax lines.

MSLPF is for distributed microstrip lowpass filter design with Chebyshev and Butterworth responses.

SLLPF designs distributed stripline lowpass filters with synthesis of electrical and distributed circuits.

MSPCF performs bandpass filter synthesis of parallel coupled lines on microstrip designs.

SLPCF designs stripline distributed bandpass filters using parallel coupled lines **Microwave Software Applications, Inc., Norcross, GA. INFO/CARD #146.**

Tatum Labs

ECA-2 is an analog circuit simulator with AC, DC, transient, and fourier analysis. **Tatum Labs, Inc., Ann Arbor, MI. Please circle INFO/CARD #145.**

EEsof

Touchstone provides an interactive design environment for the analysis, simulation, and optimization of linear microwave/RF devices and circuits.

Microwave Spice is a non-linear, time-domain circuit simulation program.

E-Syn performs automatic synthesis of noise matching, interstage matching, impedance matching, and filter networks.

MiCAD is a microwave mask layout program that converts circuit files created in the simulator program to artwork.

ANACAT is a program for the calibration, measurement, management, de-embedding, and embedding of vector network analyzer data. **EEsof, Inc., Westlake Village, CA. INFO/CARD #144.**

Bishop Graphics

Quik Circuit™ is a software package for printed circuit board design and manufacturing.

Pathfinder™ coupled with *AutoCAD™* (graphics software) provides advanced PCB design. **Bishop Graphics, Westlake Village, CA. INFO/CARD #143.**

T & E Software

Circuit Analysis analyzes linear circuits up to and including 27 nodes at user defined frequencies. **T & E Software, Raytown, MO. INFO/CARD #142.**

Compact Software

Super-Compact® PC does analysis and synthesis over a large range of RF and microwave circuits performs both ladder and nodal circuit analysis.

AutoArt™ PC is an interactive drafting program for microwave circuits.

Sonata is designed for the synthesis and optimization of microwave oscillators.

Linmic+ is tailored for MMIC development and millimeter wave applications.

RF Design Kit™ performs system optimization, WB transformer synthesis, and oscillator design.

PLL Design Kit™ does VCO design, stability analysis, switching analysis, and non-linear analysis.

Communications Design Kit™ is a synthesis and analysis program for analog and digital communication circuits.

Filter Design Kit™ is a software package for the synthesis and analysis of lumped element filters, elliptical filters, helical filters, crystal filters, and interdigital filters.

Complex Match™ performs synthesis and optimization of lumped element or distributed element matching networks.

CADEC 4™ is a general purpose package that performs analysis and optimization on virtually all circuits.

Netcom is a bidirectional HP8510/8753 controller software package that allows total control of the HP network analyzer with IBM PCs with a HP IEEE bus controller card. **Compact Software, Paterson, NJ. INFO/CARD #128.**

Step Electronics

ASM-2000 and *ASM-2500* are translation programs that take the output from *AutoCAD* or *MiCAD* and converts the data



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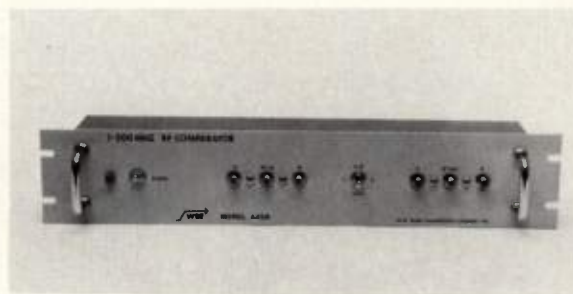
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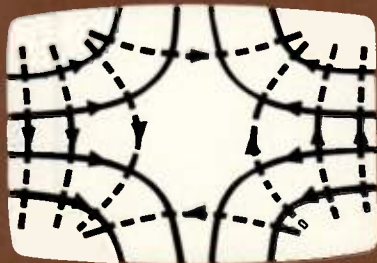
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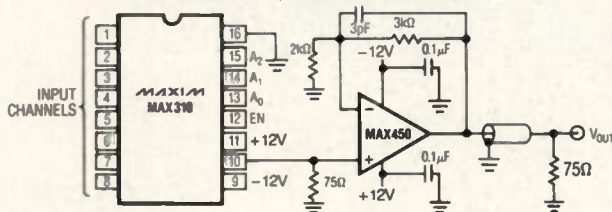
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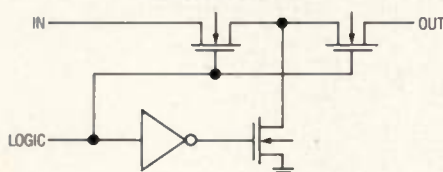
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Getting The Most Out of Programmable Instruments

Standard Interfaces and New System Software Eases the ATE Engineer's Task

By Cliff Morgan,
Tektronix, Inc.

The last 15 years have seen dramatic increases in the complexity and capabilities of test and measurement equipment. Prior to 1972, most instruments were analog, requiring a skilled operator for operation and measurement interpretation. With the advent of microprocessors, integrated A/D converters and low-cost RAM and ROM, instruments took on a digital complexion, offering immediate advantages over purely analog units. The next logical step (and current manufacturing trend) for digital instrumentation is programmability — instruments that give engineers the flexibility and speed of automatic measurements. This article describes the present state of programmable instrument use, with background on historical developments.

One of the biggest advances in digital instrumentation has been the development of standard interface buses such as IEEE-488 and RS-232. These buses provide the means for a minicomputer or programmable calculator to control and acquire data from an instrument, then process it into a more usable format. Test results can then be displayed and recorded on a printer, plotter, CRT, or magnetic media.

The electronics industry quickly adopted these interface standards, which provided the impetus to develop fully programmable test equipment. The control of instrument functions via computer ushered in the era of Automatic Test Equipment (ATE). Unfortunately, assembling a system of test equipment is often a complicated task due to variations in programming protocol. Fortunately, there are recent advances and industry trends that should minimize or accommodate these variations and ease the task of system integration.

Types of Interface Buses

Prior to the '70s there were few attempts at standardization. Generally, each manufacturer created its own special interface

to connect a particular minicomputer to a specific piece of test gear. To more fully appreciate the test system designer's task, it is helpful to review the development of several common interface buses.

One of the earliest attempts at standardization was Computer Automated Measurement and Control (CAMAC), initiated by the European scientific community primarily for nuclear instrumentation. A large controller was required, and each

signed for connecting terminals and other peripherals to mainframe computers. Since it offers no easy means of addressing instruments and since serial mode is inefficient at transferring large amounts of data, it was not intended as an instrument bus. However, most manufactures are providing optional RS-232 interfaces for test equipment, to allow access over a phone line via modem, or to use laptop computers which include an RS-232 interface

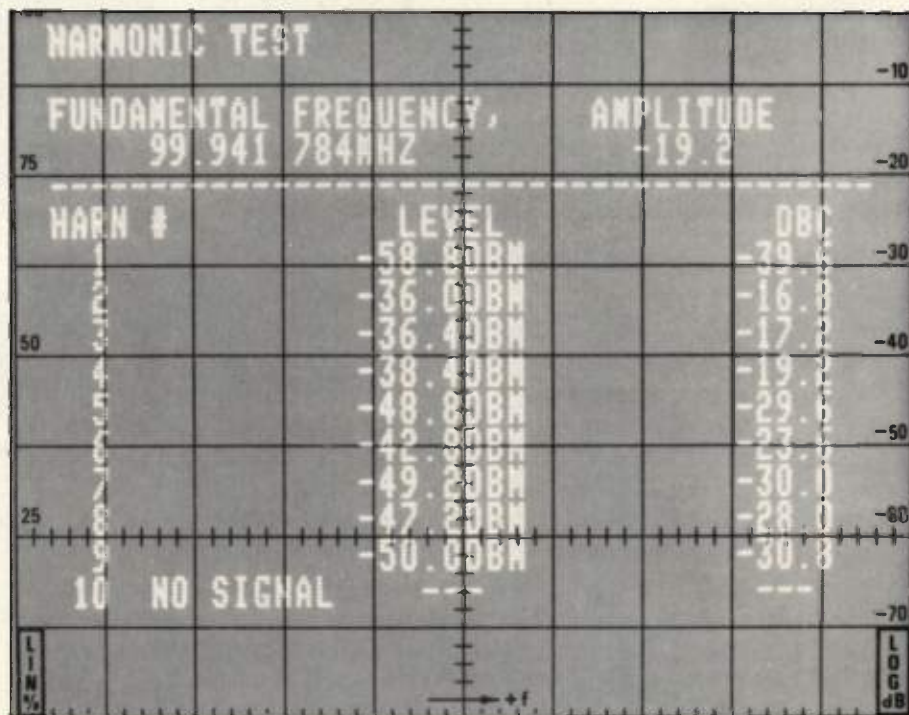


Figure 1. Sample output of a harmonic test macro.

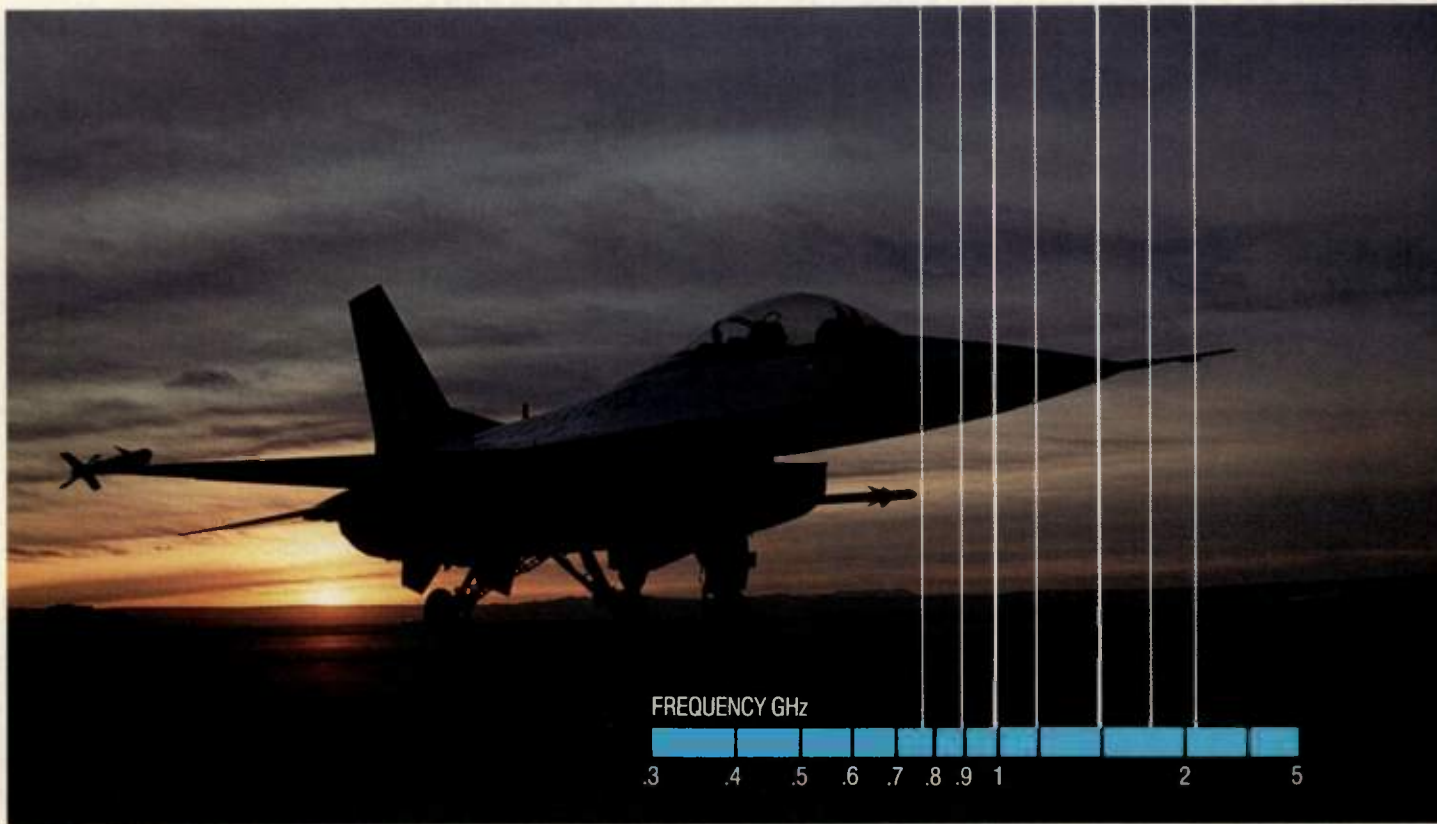
piece of test gear required a plug-in interface adapter. Although this standard was eventually accepted as IEEE Std. 583-1975, it has seen little use outside of the scientific community.

Another early interface standard, still widely used today, is the RS-232 developed by Electronics Industries Association (EIA). This is a serial interface de-

signed for connecting terminals and other peripherals to mainframe computers.

In 1972, Hewlett Packard introduced a bus to interface its programmable test equipment with an instrument controller. The HP bus was a byte-serial, bit-parallel data bus that could support a controller and one or more talkers/listeners. The concept of the Hewlett Packard Interface Bus (HP-IB) was soon endorsed by the

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electronics industry and, after some modification, was adopted in 1975 as IEEE standard 488-1975. It was later revised into the present document (IEEE 488-1978). Today it is commonly referred to as either the IEEE 488 bus, HP-IB, or General Purpose Interface Bus (GPIB). The European IEC 625 bus is identical except for the specification of bus connectors.

In spite of the many merits of the IEEE 488 standard, it did not specify the programming protocol used when one device talks or listens to another. Shortly after it was approved, Tektronix, Inc. assembled a Corporate Interface Group to try to standardize the format of commands and message protocol. The result of this committee's work was the Tek Codes and Formats document. The goal of this interface standard was to ensure that any two Tek instruments could easily communicate with each other. Tek's internal standard also served as a starting point for industry-wide efforts to standardize bus message protocol and data format.

IEEE 488.2

While industry was quick to accept the original IEEE 488 standard, the process of standardizing programming protocol has taken longer. The first industry-wide effort resulted in IEEE 728-1982, "Recommended Practice for Codes and Format Conventions for Use with ANSI/IEEE Std 488-1978." The most recent effort at language standardization has resulted in IEEE 488.2, approved in June 1987.

IEEE 488.2 can be thought of as a complement to IEEE 488. It describes standards that address the code, formats, protocol, syntax, and semantic concepts that system integrators have found useful in configuring systems. More specifically, IEEE 488.2 covers the following topics:

- Standard IEEE 488.1 subsets.
- Standard message handling protocols, including error handling.
- Unambiguous program and response message syntactic structures.
- Common commands useful in a wide range of instrument system applications.
- Standard status reporting structures.
- System configuration and synchronization protocols.

While a complete summary of IEEE 488.2 is not possible here, its usefulness can be demonstrated. Users of Tek instruments will recall the ID? query which provides the identity of the instrument queried and the firmware version number. A similar feature is included in IEEE 488.2, the *IDN? query, with the following response:

Field 1 Manufacturer: required



Figure 2. Open-site test system for radiated emissions.

Field 2 Model: required

Field 3 Serial Number:

ASCII 0 if not available

Field 4 Firmware level:

ASCII 0 if not available

Another useful command in IEEE 488.2 is the *LRN? query (similar to the Tek SET? query). This returns the complete front panel status of the instrument being queried.

MATE/CIIL

In 1976, the Air Force announced a competitive research and development program called Modular Automatic Test Equipment (MATE). The task was to study ATE problems and design an effective management concept to develop, integrate, apply, and control ATE on a total Air Force basis. The need for the MATE concept was created by the rapid proliferation of test equipment, each with its own unique programming language.

MATE is an entire test system made up of a central control computer and several connected automated instruments. Communication between devices occurs using standard interface buses such as IEEE 488. Programming of a MATE system is done using the ATLAS language, the output of which is Control Interface Intermediate Language (CIIL). Even though CIIL is compiled from ATLAS, the Air Force allows CIIL to develop independently of ATLAS. Thus CIIL can grow to accommodate new test instruments and

features.

The advantage of MATE/CIIL is that two instruments from different manufacturers (e.g. two spectrum analyzers) are programmed using the same such commands. Thus MATE/CIIL, unlike IEEE 488.2, attempts to create a universal language for test instruments.

Instrument manufacturers may take one of two approaches to creating MATE compatible instruments: 1) they can add an external Test Module Adapter (TMA) which translates the MATE/CIIL commands into the native language of the instrument, or 2) they can have the instrument's internal firmware recognize the MATE/CIIL commands. The second approach is generally preferred, but may be more difficult to implement.

An increasing trend among leading manufacturers is to make instruments multilingual. For example, the Tektronix 2756P (option 45) spectrum analyzer recognizes MATE/CIIL commands as well as Tek Codes and Formats commands. This instrument also recognizes commands from other analyzers in the Tek 490P and 2750P series family of spectrum analyzers, even if not applicable (as one designer stated "it will simply swallow commands without hiccupping").

Instrument Controllers

The system programmer's challenge is not only learning the command language of each test equipment, but also master-

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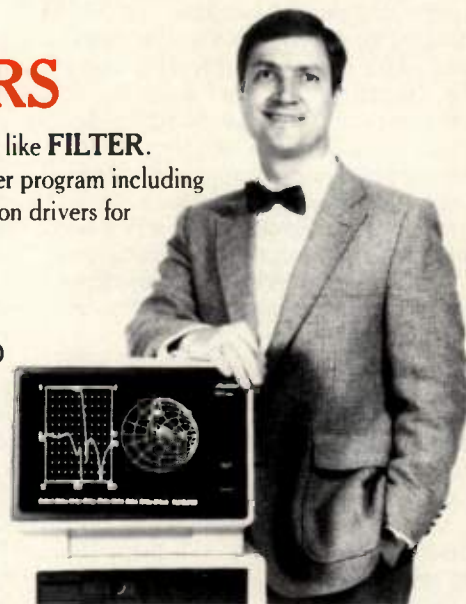
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ing the programming language of the instrument controller. These controllers fall into two broad categories: dedicated instrument controllers and personal computers.

Until recently, most ATE applications were accomplished with minicomputers or instrument controllers such as the DEC PDP-11, HP9836, or Tek 4041. They include standard interfaces and provide the necessary language features to address and control instruments and to properly handle interrupts. Another advantage is the inclusion of BASIC software for signal processing, graphics, and data reduction. These types of controllers continue to be used in both R & D and manufacturing test environments.

One drawback to the dedicated instrument controller is price. A typical controller, complete with required memory and interfaces, can cost more than \$15,000. This may not seem significant when compared to more expensive items such as network analyzers or microwave signal sources, but it may push the total system price out of reach. Another disadvantage is that the controller usually is not well suited for general purpose computing such as word processing and spreadsheet analysis, limiting its utility.

Let's contrast this situation to that of the personal computer. A fully configured PC can be purchased for under \$3,000. Also, these computers are becoming so ubiquitous that the user may already have access to a PC, eliminating the need for a special purchase. Besides low cost, PCs are supported by a wealth of software and hardware accessories. Thus data that is acquired from a programmable instrument can often be ported to a spreadsheet package for data manipulation or graphic display. With these advantages, PCs are becoming commonplace in automated test environments.

The drawback to using the PC for instrument control is that it was not designed for the application. Where a dedicated instrument controller has special commands to set and query an instrument, the same process on the PC can be cumbersome, particularly when programming is limited to BASIC.

Several manufacturers make IEEE 488/GPIB interface cards for the IBM PC™, including Capital Equipment Co., Tecmar, Ziatech, and National Instruments. When selecting a GPIB card, it is important to choose one that has a good software support (i.e. it must have device drivers for the programming language or applications software being used). For example, National Instruments' PC2 (and PC2A) card, has device drivers for several

different languages including MS-BASIC, IBM Complied BASIC 2, Turbo Pascal, and Lattice C.

For those who need a tutorial introduction to programming the PC for instrument control, Tektronix sells the GURU II package. In addition to the National PC2A card and a GPIB cable, GURU II includes device drivers, diagnostic, applications software, and a detailed tutorial manual with many examples. In recent months, several manufacturers have introduced interface cards that simplify GPIB programming. IOtech's Personal 488 interface card includes a handler that passes both literals and variable data, unlike some drivers that pass only variables. This simplifies programming calls considerably.

Another example is Hewlett Packard's HP82300A which is a 488 controller/coprocessor card for the HP Vectra (IBM compatible). This card includes a MC68000 processor with 512K of RAM, and can run programs written in Rocky Mountain Basic, the language of HP9800 series controllers.

Scientific Software for the PC

Several powerful integrated software packages have been introduced which combine the functions of data acquisition, analysis, and graphics. These packages simplify the system programmer's task by including a library of routines that perform such complex functions as integration, differentiation, convolving, correlation, and FFT. A number of powerful statistical functions are also available.

One popular package is ASYST from Macmillan Software Company. This package is a fully integrated software tool that provides the user with the most commonly used data acquisition, statistical and graphical functions required in engineering and scientific applications. ASYST is available from leading instrument manufacturers who have written instrument drivers to work with the software. This package also has GPIB drivers for a number of IEEE 488 interfaces.

The Signal Processing and Display (SPD) package from Tektronix also runs on the PC with power similar to ASYST. This package contains routines for signal processing, analysis and graphical display of data. Unlike ASYST, which is a stack-oriented programming environment, SPD routines can be called from compiled IBM BASIC or Microsoft QUICK BASIC, and from Lattice C or Microsoft C.

The Trend Toward Smarter Instruments

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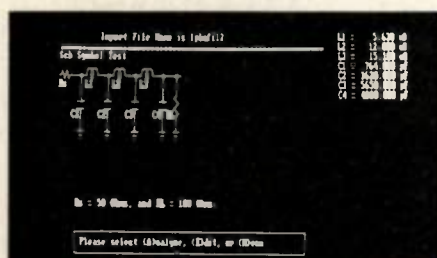
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grammable instruments had very little built-in intelligence. This meant that the system programmer had to acquire a complete array of data from the instrument, then do all of the waveform processing within the controlling computer. This not only placed a heavier workload on the programmer and instrument controller, but it also consumed considerable bus time in transferring data to and from the instrument(s).

Manufacturers are now including powerful decision-making routines within the test instrument's firmware. This innate intelligence usually involves either signal processing routines or macro programming capability. A spectrum analyzer is a good example of an instrument that benefits from signal processing. A typical application is a signal search routine which catalogs all spectral lines above a user-defined threshold. Previously, it was necessary to acquire the complete spectrum from the analyzer then do the peak find routine in an external computer.

Another significant trend in instrument firmware is the ability to create small programs called macro-functions, that can be downloaded from a controller to an in-



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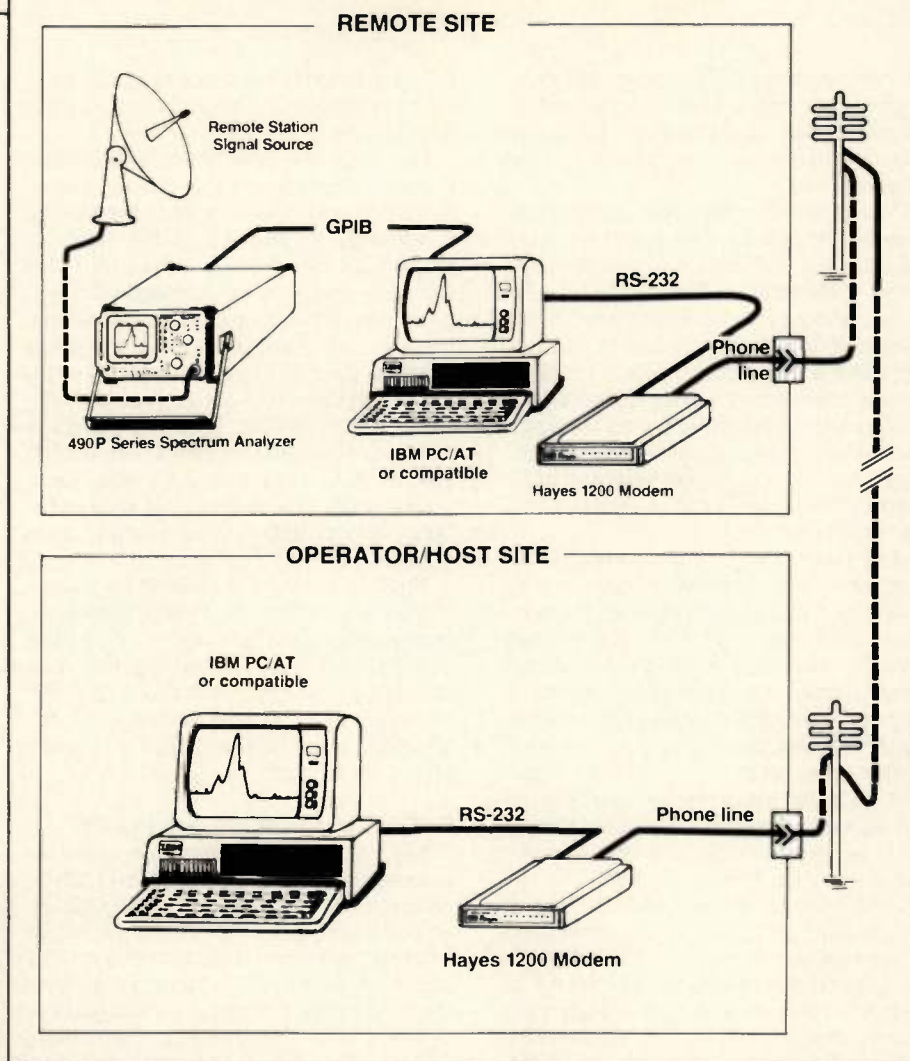


Figure 3. Equipment setup for remote-site signal monitoring.

struments' non-volatile memory. These macros can perform a complete measurement sequence independent of the controller, such as THD, S/N ratio, or spectrum surveillance. Measurement results can then be displayed on screen (Fig. 1).

Macros are particularly useful in portable instruments (such as spectrum analyzers) where it may not be convenient to haul around an external computer/controller. In this case, the macro is stored for field execution, selected from a menu displayed on the instrument CRT. Macros can also be used with external controller, to execute a particular sequence while the controller goes on to another task. Later, the controller can query the macro readout buffer to get the results.

Turnkey Systems

Another trend among test equipment manufacturers is to offer turnkey systems configured for a particular application. Such systems contain the instruments, controller, and applications software

designed for the hardware supplied. One advantage of a turnkey system is reduced programming time for the user because the software has already been designed for the system. Also, since components are supplied by a single vendor, system incompatibilities are minimized.

One application where turnkey systems are becoming quite popular is EMC (electromagnetic compatibility) compliance testing. A typical system includes a spectrum analyzer, instrument controller/computer, antennas, probes or other interference sensors, and other supporting equipment such as a preamplifier and printer/plotter (Fig. 2).

The software for EMC radiated and conducted emissions testing sets the correct frequency range, acquires raw spectral data, converts the data into proper form, and provides analysis of the data for the appropriate FCC, VDE or MIL STD specifications. Where manual EMC compliance testing can take hours, automated testing can decrease the testing time to

a few minutes, with greatly reduced operator error.

Another novel application for programmable instruments is remote-site monitoring of a signal source, such as a radar station. Tektronix' new Remote Site Monitoring (RSM) software package allows control of virtually any Tek programmable instrument using PC's and phone lines (Fig. 3). The PC is located at the remote installation, and is connected to a spectrum analyzer and perhaps other programmable instruments via the GPIB. The host-site PC is located at some convenient spot, perhaps an office or lab, where the user also has complete control over the instruments. The Tek RSM package also provides a mode for automated signal monitoring. In this mode, the remote site spectrum analyzer can continuously determine whether the signal falls within a user-defined amplitude and frequency window, and provide a warning when the signal falls out of tolerance.

Summary

These are only a few applications of programmable instruments. Nearly every manual test procedure for components and systems can now be performed faster and more accurately using the programmability of instruments now on the market. While these instruments are more expensive than their non-programmable counterparts, the cost is easily justified when the same test must be performed repeatedly. The factors that are most significant to automated test system development are:

- 1) A variety of interface options now exist (IEEE 488, RS-232, HP-IL, etc.).
- 2) Emergence of common programming protocol or bus languages for instruments (IEEE 488.2 and MATE/CIIL).
- 3) Availability of programming tools for those who are using the PC as a system controller (GURU II, ASYST, etc.).
- 4) Instruments are increasingly including intelligent signal processing routines, easing the burden on the programmer and instrument controller.
- 5) Decision-making capability, in the form of downloadable macros, allows more efficient use of the instrument controller and IEEE 488 bus.
- 6) Applications assistance, in the form of factory applications engineers and users groups, is available.

As with any evolving engineering technique, there are still obstacles to be overcome when configuring and programming test equipment for a particular application. With recent industry trends, however, the task is becoming much easier and the benefits much greater. rf

About the Author

Cliff Morgan is applications engineer at Tektronix, Inc., Frequency Domain Instruments Div., P.O. Box 500, Beaverton, OR 97077. He can be reached at (503) 627-7111.

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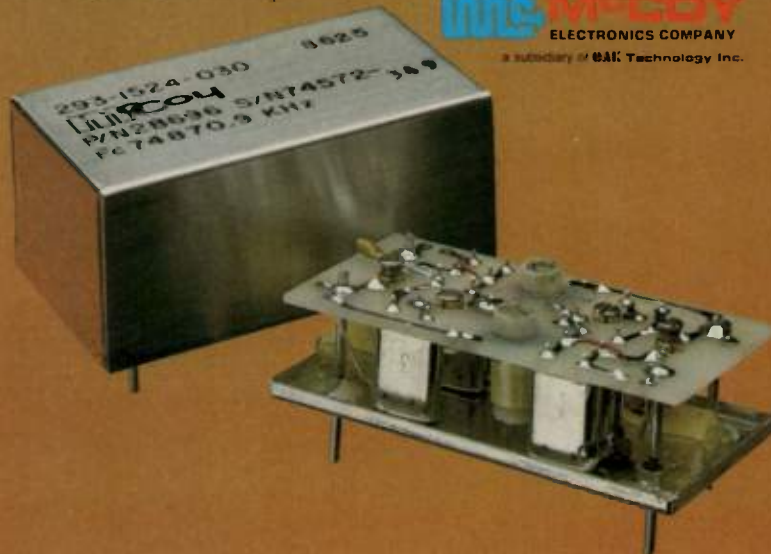
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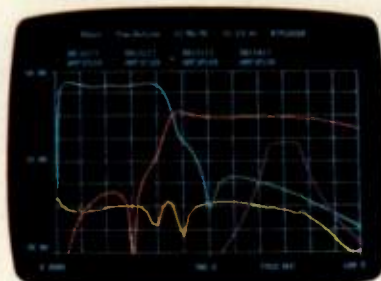
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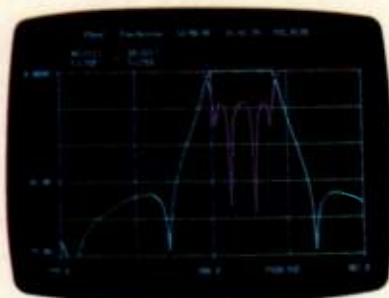
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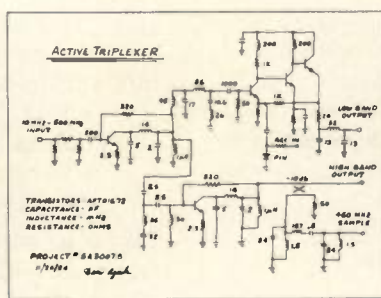
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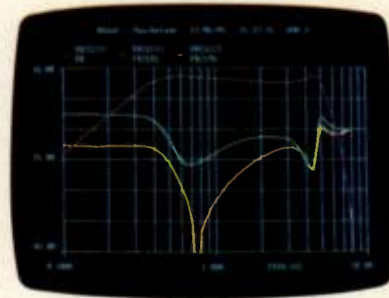
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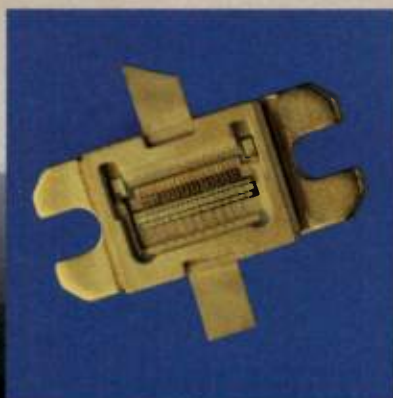
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AM 0814-175	0.85-1.4	175	7.7	52	35	75
AM 1214-175	1.2-1.4	180	8.0	55	40	75

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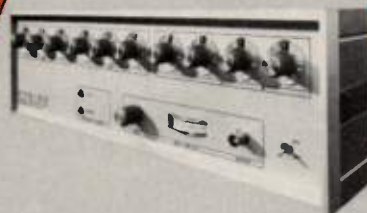
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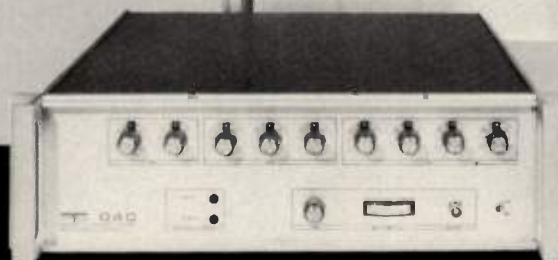
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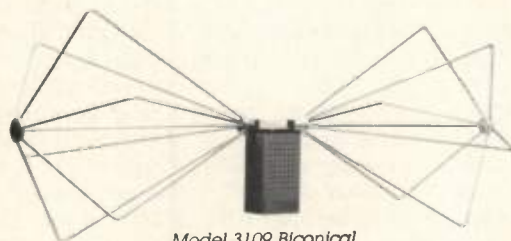
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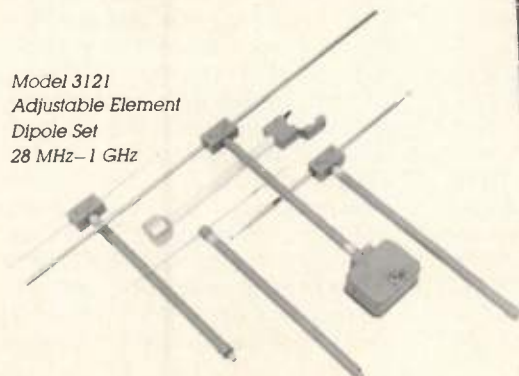
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EMC Test Labs: Standards and Requirements

***A Qualified EMC Test Laboratory and Personnel
are Valuable Assets to Today's Electronics Industry.***

By Mike Howard
Norand Corporation EMC Test Lab

EMI/EMC test labs are becoming an indispensable service to the electronics community. This necessity has been generated by voluntary and regulatory EMI test requirements for a broad range of electronic products in the commercial and military market place. The scope of an EMI/EMC lab and its organization is as varied as the range of test standards and requirements that exist. The type of EMC lab required for a selected product may range from a materials or component evaluation EMC lab to a full scale EMC system lab for evaluation of the electromagnetic compatibility of large racks of equipment to large scale vehicles.

This article will detail items an individual or organization contemplating their first EMC test and/or improvements and enhancements needed for an existing EMC lab. Measurements for EMI/EMC range from the simplistic to the very complex in form and requirements. This article will demonstrate the need for laboratory accreditation such as the EMC NVLAP (National Voluntary Accreditation Program) sponsored by the NBS (National Bureau of Standards). Personnel certification is a needed requirement as well, but to date, no programs exist for EMC personnel certification.

At certain points in the developmental phase of a product for compliance to an EMC standard(s), four basic EMC test facilities services are required. They are as follows:

I. Materials and Component Level Analysis

At the start of any design the selection of materials and components must be finalized. These items must be selected with care to maintain the maximum performance versus cost. The EMC test lab can provide such test services as shielding effectiveness tests of conductive paints and plastics used as RF barriers. Insertion loss measurements of line filters,

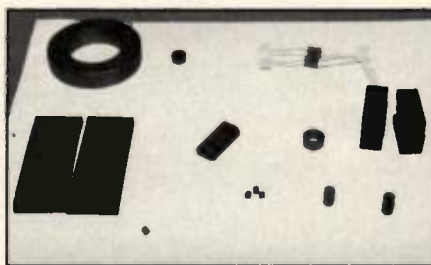


Figure 1. A variety of components can be evaluated at bench level.

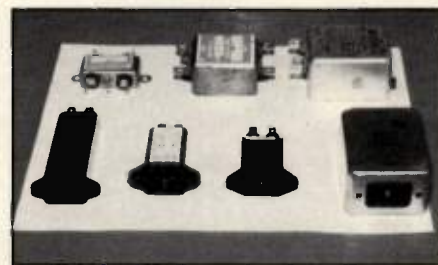
ferrite beads, etc., can also be performed to aid the design engineer in selecting the optimum materials and components for his design. Figure 1 illustrates components that can be tested at bench level.

II. Prototype and Sub-system Tests

These tests should be performed as early as possible in the product design build schedule to ensure that it is on track for compliance with the requested EMI/EMC standard. If major changes in EMI hardening is required, it is much easier to implement these changes at this point in the design and testing phase. Again, bench level tests may be used for evaluation of the product at this stage, or testing within a shielded anechoic chamber or an open-field test site may be required. Such items as raw PCB's, power supplies, and sub-assemblies can be evaluated individually to characterize and harden their EMI/EMC performance before they are integrated into the final design.

III. Production and Full System Level Tests

In essence, these should be the final verification of compliance of the product to the requested EMI/EMC test standard. If non-compliance is detected in these tests, then only minor design changes should be needed if proper testing has occurred during the materials, components, and prototype stage. Major EMI/EMC design issues must be shaken out in these early stages to minimize major EMI



design changes at the tail end of the product development cycle. EMI/EMC design considerations must be considered, tested, and proven up front. A wide range of EMC test facilities may be required for this final production or system EMC evaluation.

IV. Audit Testing

This is a must if even only on a voluntary basis versus mandatory audit test requirements. Minor or insignificant changes in the design may compromise the product's compliance with a requested EMI/EMC standard. Such changes as PCB layout, parts selection, process changes and many others can alter compliance. All design changes should be reviewed, with the EMC test lab staff being included, to determine if and what testing will be required. As a minimum, a product should be audited on a yearly basis. Some degree of statistical testing may also be required. For example, in VDE 0871 testing for West Germany, if a production product is found to be over the limit by no more than 2dB, the production run is still acceptable if it can be proven that a minimum of 80% of the equipment in the production run does not exceed the limit. A minimum sample of five units, up to a maximum of twelve units, may be required to demonstrate this requirement.

Experience needs to be, and is, developed with these four levels of test evalua-

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SAS-200-512	200 - 1800 MHz	Log Periodic	SAS-200-560	per MIL-STD-461	Loop - Emission
SAS-200-518	1000 - 18000 MHz	Log Periodic	SAS-200-561	per MIL-STD-461	Loop - Radiating
SAS-200-530	150 - 550 MHz	Broadband Dipole	BCP-200-510	20 Hz - 1 MHz	LF Current Probe
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tion by the EMC lab and the design engineers of the submitted product. This is to ensure compliance of the product and to develop a resource and data base for future designs to draw upon. The percentage of first time designs found non-compliant as submitted by design personnel with a minimum of first time hardening experience is quite high.

Several key items must be in place to ensure having a competent EMC test lab

and personnel to perform measurements in the four areas addressed earlier. They are as follows:

Support Facilities

Consideration must be given to adequate facilities to house and operate test mediums from bench level to large anechoic chambers and open-field test sites. Planning for future expansion is critical here. Many EMC test labs are ex-

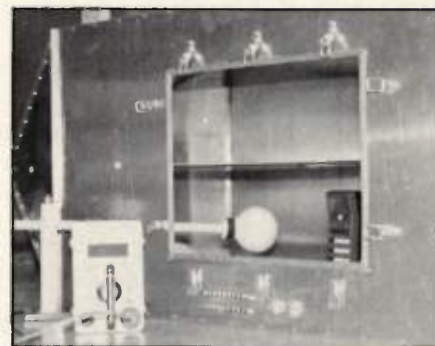


Figure 2. A TEM cell is a versatile evaluation test.

periencing a doubling or tripling effect in their operations in a few short years. This is due in part to the increased awareness of EMI/EMC standards by Industry today and recent regulatory enforcement of regulations such as the FCC rulings on computing devices (i.e., FCC Part 15 Subpart J). The support facilities must give considerations to such items as:

- areas for bench level tests
- pre-stage area for demonstration, repair, and set up of products to be tested
- storage areas for controlled samples, compliance EUT samples, and support equipment
- office space for test personnel and product engineers in attendance at the lab.
- specialized test mediums such as large anechoic chambers and open-field sites. These test mediums require large floor space and land requirements as well as specialized support in electrical, heating, cooling, and fire protection. Compliance with facility standards specified in such documents as the FCC's MP-4 and OST-55 and the NBS NVLAP standards must be met.

Test Facility

As discussed earlier, the actual test facility can range in size and complexity depending on the scale of the testing required. Some of the test facilities commonly utilized by EMC laboratories today are described below.

Shielded Chambers: A shielded chamber is typically constructed of single or dual walled metal plates bonded mechanically or their seams welded continuously to provide an RF tight seal. These shielded rooms are well suited for both emissions and susceptibility tests. The shielded enclosures do have a deficiency in that they can introduce serious measurement errors as large as +40 dB. The high conductivity and reflectivity of the enclosure walls set up standing waves that may interfere with the signal being



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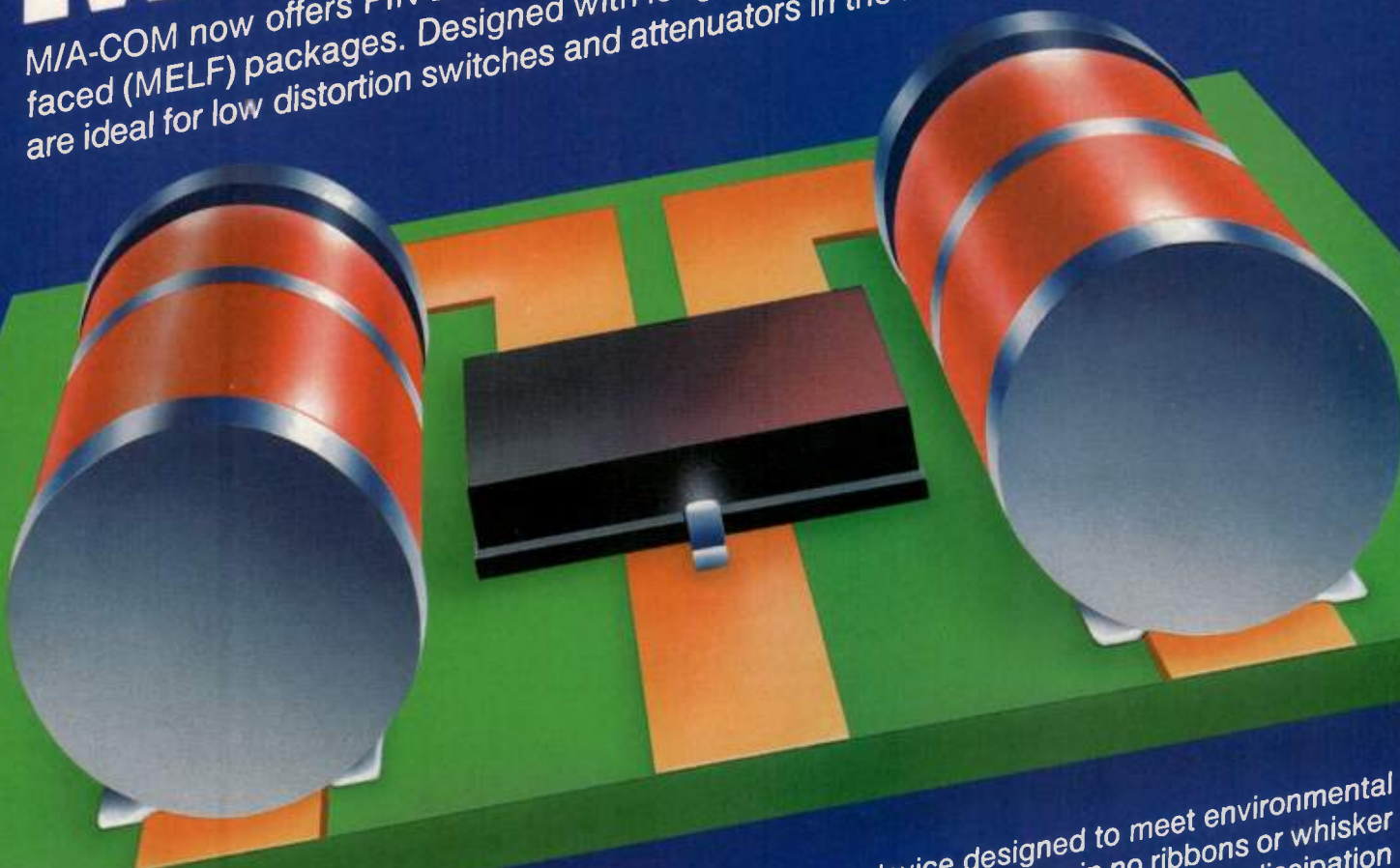
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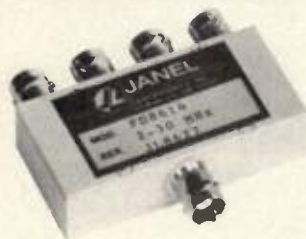


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measured and cause large gradients in the susceptibility test field. Absorbing material can be used to suppress these reflections but this may only be effective above 200 MHz. Below 200 MHz, alternate techniques being explored today are the reverberation chamber using mode-stirred or mode-tuned methods of reverberation to suppress the unwanted TEM modes within a shielded enclosure.

TEM Cell: The TEM Cell is a 50 ohm triplate transmission line with its sides closed. This design prevents radiation of RF energy into the environment and provides electrical isolation. It can be used for both emission and susceptibility measurements. The TEM Cells have seen a growth in usage due to its versatility, measurement accuracy and ease of operation. The TEM Cell is limited in the physical size of product that can be evaluated and has an upper useful frequency range determined by the cell's physical size and geometry. The TEM cell can be used for both radiated emissions and susceptibility testing. Some shielding effectiveness tests can be performed with a dual TEM cell arrangement. See Figure 2.

Open Field Sites: An open field site is just the opposite of the shielded enclosure type of facility. An open field site provides both a reference and a perturbation free environment in which actual operational conditions of the equipment can be simulated. Open field sites must cope with the environment of the RF ambient that is present. Facility personnel must take extra care to discern the radiated emissions from a product from those of ambient RF signals present. The open-field site is predominantly used for radiated emission measurements as specified by FCC Part 15-J and West Germany's VDE 0871. Radiated susceptibility tests can also be performed on an open-field site but is severely limited by frequencies and RF power that are licensed for testing purposes. A properly designed and run open-field site can provide very accurate measurement results. Figure 3 represents an open field test facility.

There are other types of facilities or slight variations of the ones mentioned above. But by using these three basic types of facilities in combination, the overall capability of the EMC test lab and its ability to meet future test requirements of product submitted to the test facility will be greatly enhanced.

Instrumentation

As with test facilities, a wide range of instrumentation is required to complement an EMC lab's capabilities. With

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*Patent No. #3,735,269



these different types of test facilities, each type of instrumentation will have a certain degree of measurement deficiency or a more suitable measurement capability than the other. For both emissions and susceptibility testing, more than one type of instrumentation and test method is required to be utilized. As with test facilities, the EMC Lab has it in its best interest to select the best from both worlds. For ex-

ample in emission testing, it is highly beneficial for the EMC lab to have at its disposal the use of both tuned receivers and spectrum analysers to perform its emission measurements. For susceptibility testing both discrete and swept frequency measurements need to be utilized to fully characterize the EUT's performance accurately and efficiently by the test lab.

Automation

To maintain a high degree of accuracy and throughput in the EMC lab some degree of automation is required. Due to the large number of operations required for various EMI/EMC tests, automation has shown to be a very cost effective approach in EMC testing. An overall reduction of 50 to 70 percent in test execution times can be achieved with automation of EMC tests versus manual methods. Reliable and meaningful automated EMC tests can be achieved only when a thorough knowledge of the test facilities, instrumentation, and test standards is known. Only then will the automated EMC test lab system or facility be an effective tool and aid for the EMC lab. No matter how much automation is placed within an EMC test lab, a degree of manual methods will still be needed and required particularly in the case of I/O cabling placement and movement by lab personnel to maximize emissions for test performed according to FCC Part 15 Subpart J for computing devices.

Personnel

Even with the increased awareness of EMI/EMC in the Electronics industry today, there is a critical shortage of qualified EMC engineers and technicians.

NBS NVLAP Laboratory Accreditation Program

The NBS NVLAP is a voluntary system for accrediting laboratories found competent to perform specific testing operations. NVLAP's goals are to:

1. Provide national recognition for competent laboratories
2. Provide a laboratory with a quality assurance check
3. Identify competent laboratories for regulatory agencies purchasing authorities, and product certification systems.
4. Provide guidance from technical experts to aid laboratories in reaching a higher level of performance to improve engineering and product information.
5. Foster international acceptance of test data produced by U.S. Laboratories.

The NBS NVLAP program for EMC Labs currently offers accreditation for test methods under FCC Part 15 Subpart J for computing devices and Part 68 for telecommunications terminal equipment. Additional test methods for accreditation have just recently been announced by the NBS. They include FCC Part 15 Subparts C, D, E, & H for low power communications devices and receivers, FCC Part 18 for industrial, scientific, and medical equipment, FCC Part 90 for type acceptance of radio transmitters and receivers.

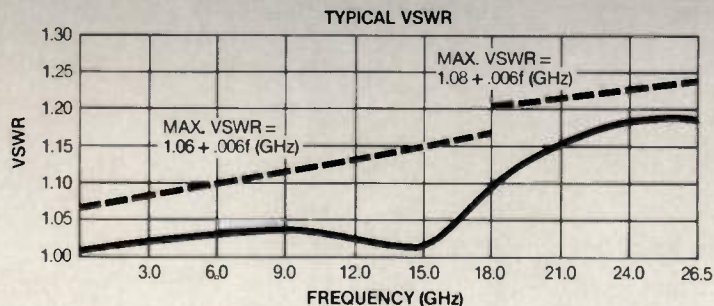
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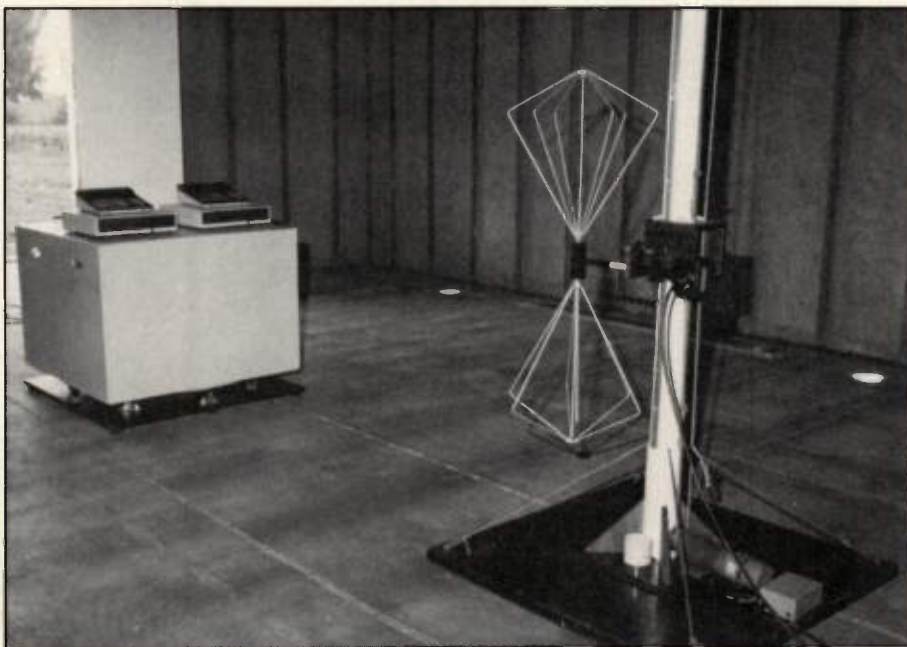



Figure 3. An all-weather enclosed open-field test site is suitable for testing for FCC Part 15-J and VDE 0871.

Selected methods under MIL-STD 462 will also be offered. In essence, the NVLAP program provides the needed qualification, standards, and quality assurance programs required for EMC Labs in an industry with a wealth of standards for product qualification and but with limited standards for facilities and personnel.

Summary

If you are considering setting up an EMC Test Lab for the first time or considering expanding or improving a present facility, careful planning and thought must be pursued to ensure that the lab will meet its intended goals today and into the future. EMC Labs will continue to see increased growth and opportunities in a wide range of areas from measurement methodology to types of facilities available to perform EMI/EMC measurements within. There will be more and improved standards established for the EMC test lab and it's personnel in the near future. It will be imperative for the EMC test lab to meet these standards established for their facility and personnel before attempting to test EMI/EMC standards of product submitted to their lab. The undertaking and operation of an EMC test lab is in no way a light task. Commitments in facilities, capital equipment and qualified personnel must be set in a firm foundation to provide the quality and quantity that is expected from an EMC Test Lab. A basic EMC test lab will typically cost from \$100,000, with a staff of one,

to as large as \$10,000,000 in investment, with staffs as large as 30 personnel. A new and exciting future lays ahead for today's EMC Test Labs. 

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3. Armfield, T., Howard, M., and Walter, S., "Automated Component Level EMC Test Facility", IEEE EMC Symposium Record, 1982, pp 364-369.
4. NBS NVLAP Electromagnetics LAP Handbook, Oct. 1985.
5. Robertson, Richard G., "General Overview of Automated EMI Measurement Systems", EMC Technology, January 1982, pp 16.

About the Author

Mike Howard is the supervisor at the Norand Corporation EMC Test Lab. He has over 15 years experience in various types of commercial/military EMI/EMC test and design. He can be reached at Norand Corporation, EMC Test Lab, 550 2nd St. S.E., Cedar Rapids, IA 52401. Tel: (319) 846-2415.

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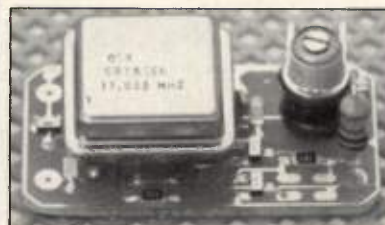
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PIN Diode Switches — Part III

Transmit/Receive Switch Applications

By Andrzej B. Przedpelski
A.R.F. Products, Inc.

The PIN diode T/R switch is described in this article. Shunt and series SPST switches were described in Parts I (RF Design, Nov. 1986) and II (RF Design, Feb. 1987). Combining several switch sections can provide other serviceable switch configurations.

A T/R switch is a SPDT switch used to switch a single antenna between a transmitter and a receiver. This can be done with two or more PIN diode switches, depending on system requirements. As an illustration, the following requirements are specified:

- minimum transmitter path insertion loss
- high attenuation in receiver path
- low overall power consumption
- receiver ON period is 10 percent of total time
- simple diode switching drive

To provide a simple switching drive circuit, it is desirable to use the same drive for the receiver and the transmitter switch. For minimum power consumption, the diodes should be back biased during the transmitting period since the transmit period occurs 90 percent of the time. A simple SPDT switch using two series diodes (one in each leg) will not satisfy these requirements. However, using one shunt switch and one series switch configuration satisfies both.

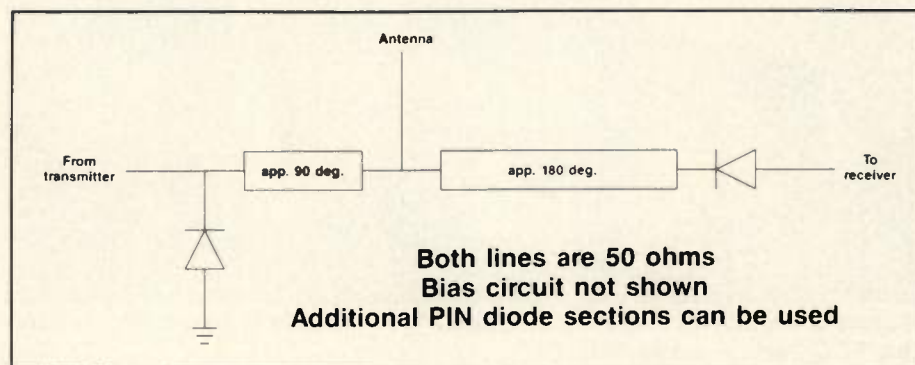


Figure 1. Simple T/R switch configuration.

uration satisfies both.

It is more desirable to use the shunt configuration in the transmitter path and the series configuration in the receiver path, as shown in Figure 1. In this configuration, the transmitter diode is out of the circuit (back biased) during the transmit period. Hence, it is not required in order to pass the large RF current. Its only requirement is to withstand the combined DC-bias and peak RF voltage and a low power diode can still be used. Insertion loss of this configuration is low, as described in Part I.

In the receiver path, the isolation can be increased by using several diodes in series, as described in Part II. Figure 2 shows the equivalent circuits in the transmit and receive modes. In the transmit mode, the transmitter diode is in its high

impedance state, and its effect on the transmitted signal is usually negligible (its reactance can be compensated by transmitter tuning). The receiver diode appears directly at the antenna terminal (a 180 degree line does not change the impedance) as a high series impedance. In the receiver mode, the receiver diode places a small resistance in the receive path, introducing some loss. The transmitter diode has also a low impedance, which is transformed by the 90 degree line into a high impedance at the junction. Switching the antenna between the transmitter and the receiver is then accomplished.

The transmitter 90 degree line is obviously needed to provide the impedance transformation shown in Figure 2. There is no need for the receiver 180 degree line. It is used for convenience to adjust

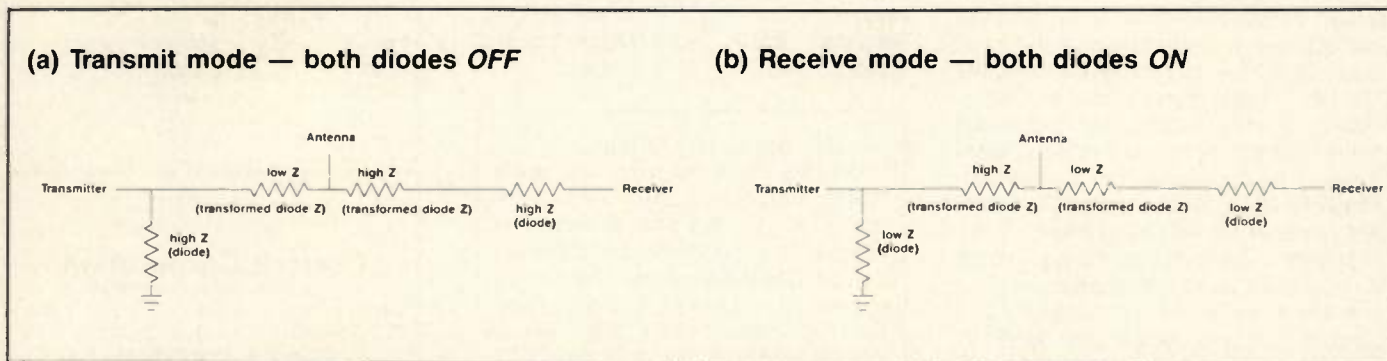


Figure 2. Equivalent circuit.


Angle	R	X
80	62.943	268.97
81	76.642	295.95
82	95.265	328.18
83	121.38	367.07
84	159.35	414.27
85	216.94	471.23
86	308.34	537.1
87	459.16	601.3
88	706.69	618.78
89	1046.5	461.07
90	1250	6.2832
91	1054.2	-453.89
92	713.77	-617.81
93	463.64	-602.52
94	311.03	-538.68
95	218.6	-472.66
96	160.42	-415.47
97	122.1	-368.05
98	95.771	-328.99
99	77.009	-296.63
100	63.216	-269.54

Table 2. Receiver OFF transferred impedance.

Angle	R	X
170	266.25	892.33
171	551.8	1222.8
172	1510	1627.7
173	3253.2	-199.84
174	1336.9	-1605.3
175	504.12	-1179
176	249.21	-865.67
177	146.35	-673.88
178	95.781	-548.79
179	67.425	-461.62
180	50	-397.89
181	38.55	-349.18
182	300.633	-310.83
183	24.934	-279.86
184	20.698	-254.32
185	17.465	-232.9
186	14.941	-214.66
187	12.934	-198.95
188	11.312	-185.26
189	9.9825	-173.22
190	8.8791	-162.55

Table 1. Transmitter OFF transferred impedance.

the impedance at the junction to any desired impedance at the switch junction. It also compensates for the diode impedance in the OFF position. It is apparent that the two transmission lines perform useful functions in the respective diode OFF conditions. This can be demonstrated by using typical PIN diodes with an OFF capacity of 0.2 pF. The operating frequency is 2000 MHz.

Table 1 shows the transmitter diode OFF impedance, transferred to the function point. It appears that a 90 degree line would be most suitable. Table 2 shows the receiver diode OFF impedance transferred to the junction point. In this case, a shorter line (173 degrees) seems preferable to the nominal 180 degrees. 

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1. SDI, N. Billerica, MA; Bulletin D-1008.
2. Hewlett-Packard; Application Notes 922, 929, 957, 971, 985.
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4. Gerald Hiller, "Designing with PIN Diodes," *RF Design*, March/April 1979 and May/June 1979.

About the Author

Andrzej B. Przedpelski is vice president, development of A.R.F. Products, Inc., 2559 75th St., Boulder, Colo. 80301. he serves as consulting editor to *RF Design*.

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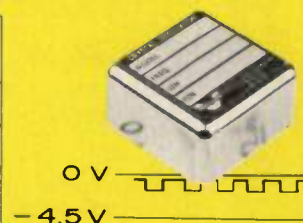
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A Low Cost 100 MHz to 2.5 GHz Amplifier

Contest entry is an excellent example of a practical engineering problem.

By David L. Tharp
Colorado Data Systems

Using low cost MMIC amplifiers in a feedback network that achieves a flat frequency response up to 2.5 GHz earned the author a copy of Superstar 3.0 in the RF Design Awards Contest. The amplifier was developed and incorporated in a smart frequency counter board.

An amplifier with a 24 dB gain from 100 MHz to 2.5 GHz is the basic requirement of the frequency counter board. A simple approach was to utilize a drop-in 50 ohm modular amplifier. This approach, though convenient can prove to be costly. Another method is to design a system with monolithic RF amplifiers. The second technique was used and the finished product proved to be very cost effective.

The amplifiers incorporated in the design were Avantek's MSA-0204. S-parameter files for use by Touchstone™ were generated by extrapolating data from the MSA series spec sheets (taking antilogos where necessary). A hypothetical single-stage amplifier was then constructed in a circuit file for experimentation. This is shown in Figure 1. Various devices were tried and it was observed that the amplifier required several low gain stages since

the high gain MSA devices rolled off too quickly. Also, some kind of additional gain equalization was required.

Figures 2 and 3 illustrate that capacitive feedback can be used to achieve the needed equalization. Rather than calculate the values for the feedback network, a resistor and capacitor node was added to the circuit file. Variable values with constraints placed on practical sizes were specified before allowing Touchstone to optimize the circuit. Figure 4 illustrates that Touchstone came up with values that equalized the gain to about 6.8 dB ± 0.25 dB across the required bandwidth.

A circuit file simulating four of the single stage amplifiers was implemented using the S-parameter output of the single stage simulation (Figure 5). The frequency response and the return loss of the circuit met the design goals. This is illustrated in Figures 6 and 7.

A prototype was constructed on a sample of 0.015" Rogers 5870 material. This resulted in a 50 ohm microstrip width of 0.044". This closely matched the width of the interstage coupling capacitors and the lead width of the MSA devices. The

recommended bias for the MSA-0204 is 5 V at 25 mA. Since the supply voltage is 7.5 V, a bias resistance of 100 ohms was required for each stage.

In spite of less than ideal construction methods, particularly in the feedback networks, the performance of the prototype circuit had a high degree of correlation between the theoretical and measured results. Figure 8 is a plot of data as measured on a spectrum analyzer and a plot of the theoretical result as outlined by Touchstone.

This circuit is an example of what can be done in a short period of time with modern RF components and design tools. Exclusive of the PC board and artwork, the entire circuit was constructed for under \$15 in parts. □

About the Author

David L. Tharp is an RF design engineer at Colorado Data Systems, 3301 West Hampden Avenue, Englewood, CO 80110. He can be reached by telephone at (303) 762-1640.

```

CIRCUIT FILE: "AVANT204.CKT" SINGLE STAGE AMPLIFIER USING
DATE/AUTHOR: 1/8/87 -- DAVE THARP
VAR
FEEDR # 1000 272.43880 10
CCAP # 1000 849.35680 10
LWID = .044
ICAP = 100
DIM
FREQ GHZ
RES OH
COND /OH
IND NH
CAP PF
LNG IN
TIME SEC
ANG DEG
EQN
CKT
MSUB ER=2.33 H=0.015 T=0.0014 RHO=.686 RG=0
TAND TAND=0.00012
*****
SUBSTRATE ENTRY FOR ROGERS 5870 1 OZ COPPER -- RHG IS TO BE
DETERMINED (RMS SURFACE ROUGHNESS)
TAND IS THE DIELECTRIC LOSS TANGENT FOR 5870 (10 GHZ)
*****
MLIN 3 4 W*LWID L= 0.375 ! MICROSTRIP GOING TO THE MODAMP
MTAPER 4 5 W1*LWID W2 = .020 L= .05 ! TAPER TO 0404 LEAD WIDTH OF .03
S2PA 5 6 0 MSA0204 ! MSA-0204 MODAMP 2-PORT DEVICE
MTAPER 6 7 W1 = .020 W2*LWID L= .05 ! TAPER BACK OUT TO 50 OHM
MLIN 7 8 W*LWID L = .5 ! MICROSTRIP TO DC BLOCK CAP
CAP 8 9 C*ICAP
RES 5 10 R*FEEDR
CAP 6 10 C*CCAP
DEF2P 3 9 AVANT204
DEFINE THE ABOVE AS A 2-PORT
TERM
PROC
OUT
AVANT204 MAG[S11] ! THESE ENTRIES USED TO
AVANT204 ANG[S11] ! PRODUCE A "S2P" FILE
AVANT204 MAG[S21] ! FOR USE BY OTHER
AVANT204 ANG[S21] ! TOUCHSTONE CIRCUIT
AVANT204 MAG[S12] ! FILES.
AVANT204 ANG[S12] ! UNFORTUNATELY THE "SPAR" MACRO PRODUCES DB MAG[S21]
AVANT204 MAG[S22] ! SO WE HAVE TO DO IT THIS WAY.
AVANT204 ANG[S22] !
AVANT204 DB[S21] GR1
AVANT204 DB[S11] GR2
AVANT204 DB[S22] GR2
AVANT204 ANG[S11] GR3
AVANT204 ANG[S21] GR3
AVANT204 S11 SC1
FREQ
SWEEP 0.100 3 .05 ! SWEEP FROM 100 - 3000, 100 MHZ INCREMENTS
GRID
RANGE 0 3 .5 ! HORIZONTAL AXIS 0-3000 MHZ, 500 MHZ/DIVISION
GR1 6 12 1 ! VERTICAL AXIS FOR S21 6-8 DB, .5 DB/DIVISION
GR2 -30 0 5 ! VERTICAL AXIS FOR S11,S22 -30 TO 0, .5 DB/DIVISION
GR3 -180 180 30 ! VERTICAL AXIS FOR PHASE
OPT
RANGE .1 2.5 ! OPTIMIZE OVER FULL BANDWIDTH
AVANT204 DB[S21] = 7 10
AVANT204 DB[S11] < 10
TOL

```

Figure 1. Circuit file for the amplifier.

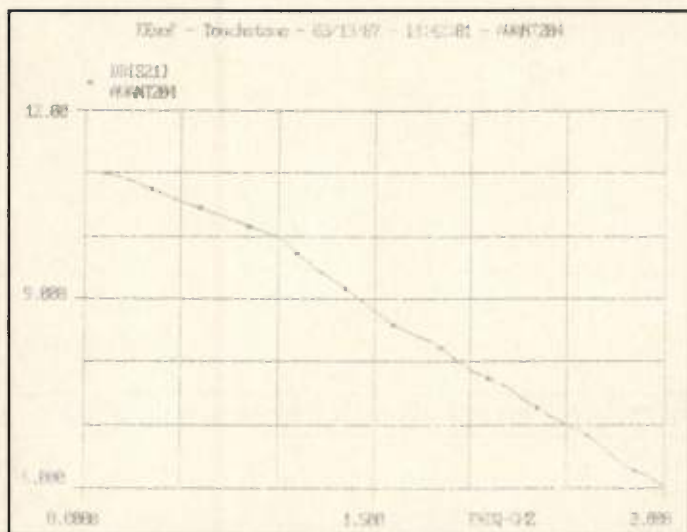


Figure 2. Plot of gain versus frequency.

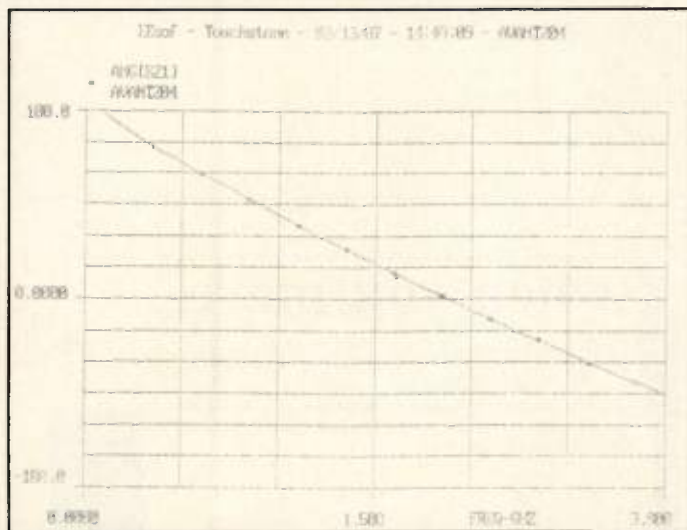


Figure 3. Plot of phase versus frequency.

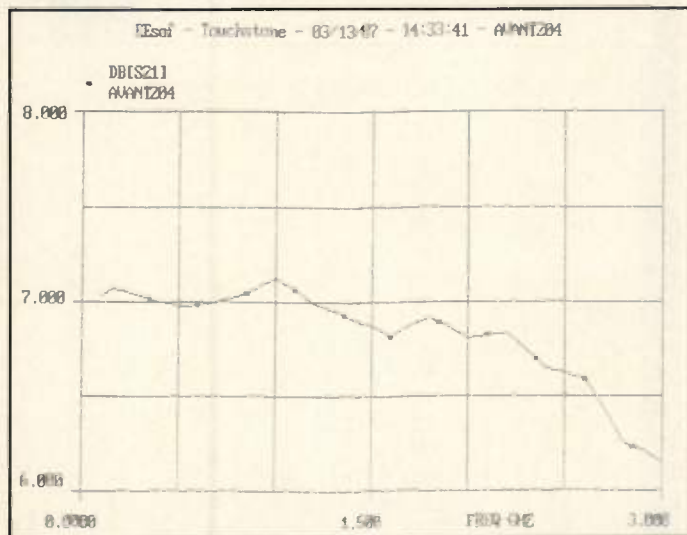


Figure 4. This plot illustrates that the equalized gain is 6.8 dB \pm 0.25 dB.

```

CIRCUIT FILENAME : MYAMP.CKT
DESCRIPTION : FOUR-STAGE AMPLIFIER USED IN 53A-561 COUNTER
              FRONT END.
DATE/AUTHOR : 8/15/86 -- DAVE THARP
              8/29/86 -- MODIFIED FOR FOUR STAGES
              9/30/86 -- MODIFIED TO USE "AVANTAMP.CKT" FOR GAIN STAGES
              11/15/86-- MODIFIED TO USE "AVANT204.CKT" FOR GAIN STAGES
VAR
ICAP = .100
LWID = .044
LONG = .375
DIM
FREQ GHZ
RES OH
COND /OH
IND NH
CAP PF
LNG IN
TIME SEC
ANG DEG
EQN
CKT
MSUB ER=2.33 H=0.015 T=0.0014 RHO=.686 RGH=0
TAND TAND=0.00012
*****
SUBSTRATE ENTRY FOR ROGERS 5870 1 OZ COPPER -- RHG IS TO BE
DETERMINED (RMS SURFACE ROUGHNESS)
TAND IS THE DIELECTRIC LOSS TANGENT FOR 5870 (10 GHZ)
*****
MLIN 1 2 W=LWID L=LONG ! INPUT MICROSTRIP
CAP 2 3 C=ICAP ! FIRST DC BLOCKING CAPACITOR
S2PA 3 4 0 A204 ! FIRST GAIN STAGE
S2PA 4 5 0 A204 ! SECOND GAIN STAGE
S2PA 5 6 0 A204 ! THIRD GAIN STAGE
S2PA 6 7 0 A204 ! FOURTH GAIN STAGE
MLIN 7 8 W=.044 L=.375 ! MICROSTRIP OUTPUT
DEF2P 1 8 FOURAMP ! DEFINE THE ABOVE AS A 2-PORT
TERM
PROC
OUT
FOURAMP DB[S21] GR1
FOURAMP DB[S11] GR2
FOURAMP DB[S22] GR3
FOURAMP ANG[S11] GR3
FOURAMP S11 SC1
FREQ
SWEEP 0.100 3.1 ! SWEEP FROM 100 - 3000, 100 MHZ INCREMENTS
GRID
RANGE 0 3.5 ! HORIZONTAL AXIS 0-3000 MHZ, 500 MHZ/DIVISION
GR1 20 30 1 ! VERTICAL AXIS FOR S21 35-45 DB, 2 DB/DIVISION
GR2 -30 0 5 ! VERTICAL AXIS FOR S11,S22 -30 TO 0, 5 DB/DIVISION
GR3 -180 180 30 ! VERTICAL AXIS FOR PHASE
OPT
TOL

```

Figure 5. Simulation of four single-stage amplifiers.

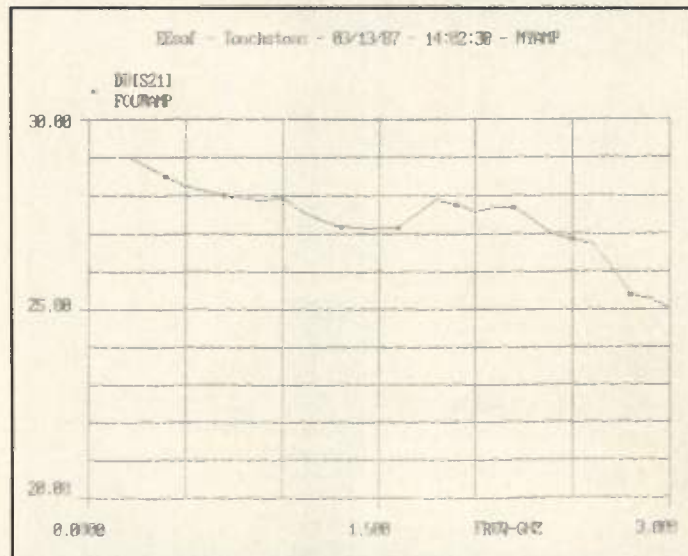


Figure 6. Plot of frequency response versus frequency

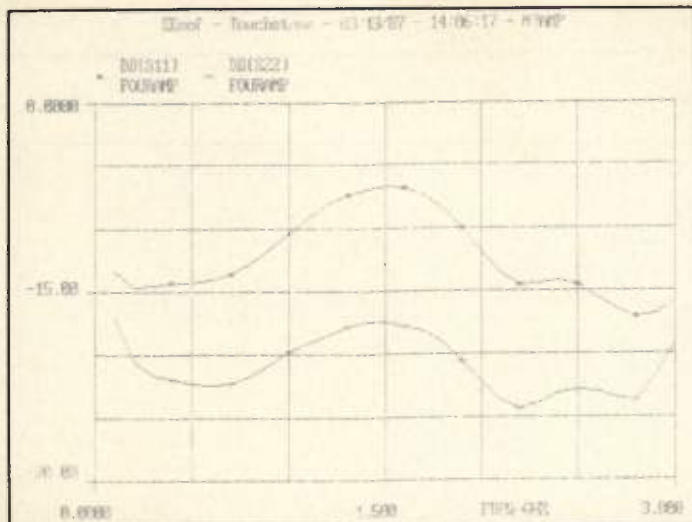


Figure 7. Plot of return loss versus frequency.

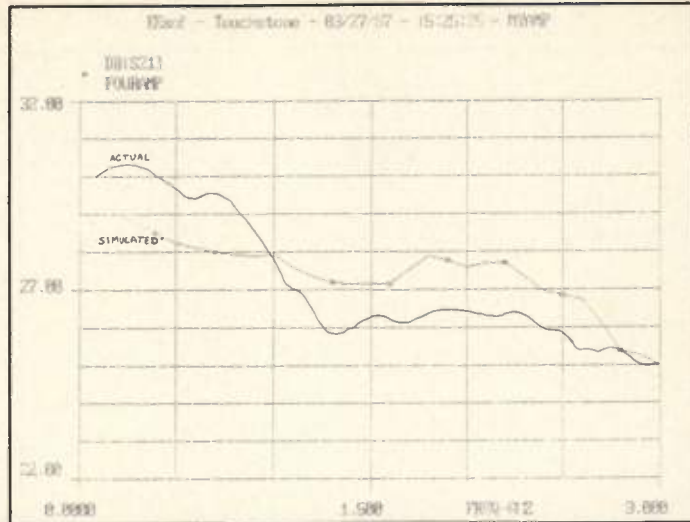


Figure 8. Actual and simulated plots of gain in dB.

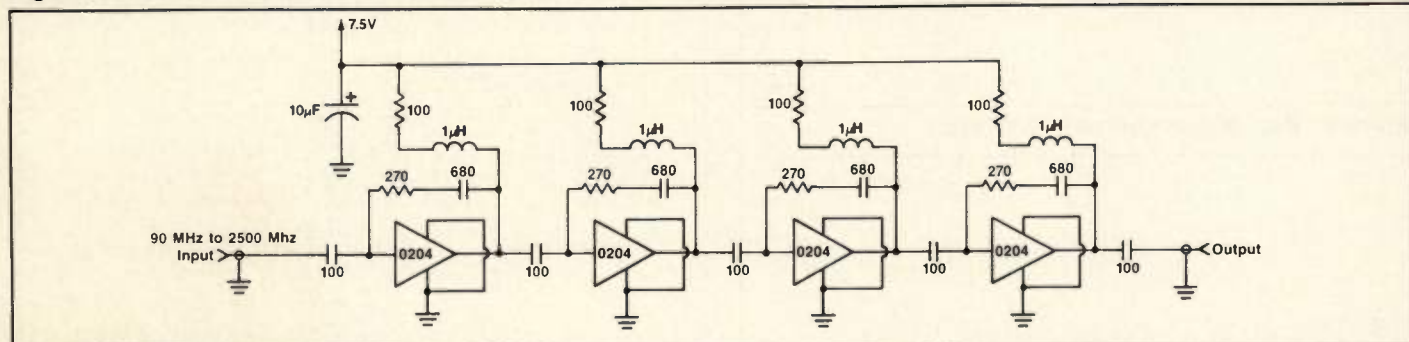


Figure 9. The final circuit.

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A Noise Specification Program

BASIC Program Converts Expressions for Noise Power

By James Mikeworth
Boeing Company

The program presented here is based on "A Noise Nomograph" by Seymour Schneider (June 1985, RF Design, p. 48). The graphical representation of that article has been replaced with a short BASIC program. To explain the principles behind the program, the following text is adapted from the 1985 article:

The output from a noise source can be expressed in many different ways, depending on the frequency range and the system with which the noise source is used. For example, at microwave frequencies the output of a noise source is typically expressed as ENR (Excess Noise Ratio), which is defined as the noise ratio minus unity, where the *noise ratio* is the ratio of the equivalent noise temperature of the noise source to room temperature.

At audio frequencies the output of a noise source is usually expressed as $\mu V/\sqrt{\text{Hz}}$, where the RMS voltage measured in a given bandwidth is divided by the square root of the measurement bandwidth.

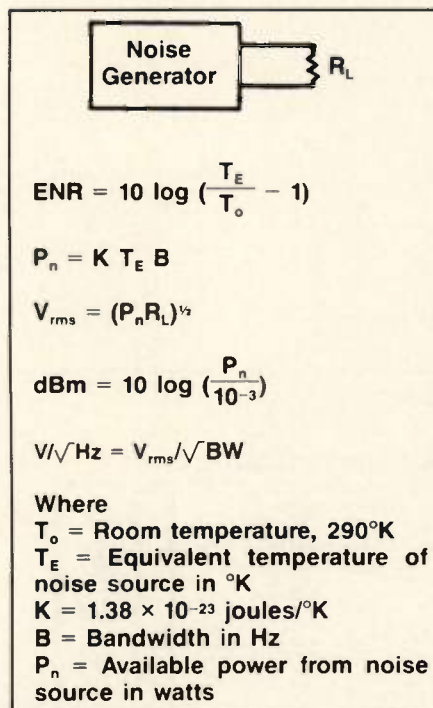


Figure 1. Noise relationships.

Other expressions are dBm (per bandwidth), V_{rms} (per bandwidth), and equivalent noise temperature. The noise program relates these commonly used terms associated with noise power calculations. In addition, it can be used to determine what effect a change in a given parameter has on noise power.

The relationships used in the program refer to a noise generator that provides power output over a specified noise bandwidth into a load (Figure 1).

This program is menu driven and will prompt the user when data needs to be entered. Figure 2 shows examples of the menu and calculations. The program covers all areas of the original Noise Nomograph, with conversions in both "forward" and "reverse" directions. As written, the program runs in IBM BASICA.

About the Author

James Mikeworth is a Circuit Design Engineer at Boeing Company in Seattle, WA. He can be reached at 11020 Kent-Kangley, D110, Kent, WA 98031.

1 ENR TO DBM	1 ENR TO DBM
2 ENR TO VRMS ACROSS LOAD RESISTOR	2 ENR TO VRMS ACROSS LOAD RESISTOR
3 ENR TO NOISE SPECTRAL DENSITY	3 ENR TO NOISE SPECTRAL DENSITY
4 DBM TO ENR	4 DBM TO ENR
5 VRMS ACROSS LOAD RESISTOR TO ENR	5 VRMS ACROSS LOAD RESISTOR TO ENR
6 NOISE SPECTRAL DENSITY TO ENR	6 NOISE SPECTRAL DENSITY TO ENR
7 EXIT TO BASIC	7 EXIT TO BASIC
ENTER NUMBER OF CHOICE? 5	ENTER NUMBER OF CHOICE? 2
VRMS ACROSS LOAD RESISTOR TO ENR	ENR TO VRMS ACROSS LOAD RESISTOR
VRMS ACROSS LOAD (μV)=? 1000	ENR=? 57 dB
RLOAD=? 50	NOISE BANDWIDTH IN MHZ=? 10
NOISE BANDWIDTH IN MHZ=? 10	RLOAD=? 50
ENR= 56.98751 DB	VRMS ACROSS LOAD RESISTOR= 1001.438 μV

Figure 2. Sample calculations, showing menu format.

```

100 REM NOISE ANALYSIS PROGRAM
110 DIM A$(7)
120 TO=290:K=1.38E-23
130 A$(1)=" ENR TO DBM"
140 A$(2)=" ENR TO VRMS ACROSS LOAD RESISTOR"
150 A$(3)=" ENR TO NOISE SPECTRAL DENSITY"
160 A$(4)=" DBM TO ENR"
170 A$(5)=" VRMS ACROSS LOAD RESISTOR TO ENR"
180 A$(6)=" NOISE SPECTRAL DENSITY TO ENR"
190 A$(7)=" EXIT TO BASIC"
200 CLS:REM CLEAR SCREEN
210 FOR I=1 TO 7
220 PRINT I;A$(I):NEXT I:PRINT:PRINT
230 INPUT"ENTER NUMBER OF CHOICE ";CN
240 IF CN(1 OR CN) 7 THEN 230
250 CLS:PRINT A$(CN):PRINT:PRINT
260 ON CN GOTO 290,360,440,520,620,730,850
270 REM
280 REM ENR TO DBM
290 GOSUB 920
300 DBM=10*LOG(PN/.001)/LOG(10)
310 PRINT"DBM=" ;DBM
320 GOSUB 870
330 GOTO 200
340 REM
350 REM ENR TO VRMS ACROSS LOAD
360 GOSUB 920
370 INPUT"RLOAD=" ;RL
380 VRMS=SQR(PN*RL)*1000000!
390 PRINT"VRMS ACROSS LOAD RESISTOR=" ;VRMS;" uV"
400 GOSUB 870
410 GOTO 200
420 REM
430 REM ENR TO NOISE SPECTRAL DENSITY
440 GOSUB 920
450 INPUT"RLOAD=" ;RL
460 V1=SQR(PN*RL)/SQR(BW)*1000000!
470 PRINT"VRMS PER ROOT HZ ACROSS ";RL;" DHMS=" ;V1;" uV"
480 GOSUB 870
490 GOTO 200
500 REM
510 REM DBM TO ENR
520 INPUT"NOISE BANDWIDTH IN MHZ=" ;BW:BW=BW*1000000!
530 INPUT"DBM=" ;DBM

```

```

540 PN=.001*10^(DBM/10)
550 TE=PN/K/BW
560 NR=10*LOG(TE/TO-1)/LOG(10)
570 PRINT"ENR=" ;NR;" DB"
580 GOSUB 870
590 GOTO 200
600 REM
610 REM VRMS ACROSS LOAD TO ENR
620 INPUT"VRMS ACROSS LOAD (uV)=" ;VRMS:VRMS=VRMS*.000001
630 INPUT"RLOAD=" ;RL
640 INPUT"NOISE BANDWIDTH IN MHZ=" ;BW:BW=BW*1000000!
650 PN=VRMS^2/RL
660 TE=PN/K/BW
670 NR=10*LOG(TE/TO-1)/LOG(10)
680 PRINT"ENR=" ;NR;" DB"
690 GOSUB 870
700 GOTO 200
710 REM
720 REM NOISE SPECTRAL DENSITY TO ENR
730 INPUT"VRMS PER ROOT HZ ACROSS LOAD (uV)=" ;V1
740 INPUT"RLOAD=" ;RL
750 INPUT"NOISE BANDWIDTH IN MHZ=" ;BW:BW=BW*1000000!
760 VRMS=V1*SQR(BW)*.000001
770 PN=VRMS^2/RL
780 TE=PN/K/BW
790 NR=10*LOG(TE/TO-1)/LOG(10)
800 PRINT"ENR=" ;NR;" DB"
810 GOSUB 870
820 GOTO 200
830 REM
840 REM EXIT TO BASIC
850 PRINT" END OF NOISE ANALYSIS PROGRAM--GOOD BYE"
860 END
870 REM WAITING SUBROUTINE
880 PRINT:PRINT"PRESS ANY KEY TO CONTINUE"
890 C$=INKEY$:IF C$="" THEN 890
900 RETURN
910 REM
920 REM STANDARD 1-3 INPUT SUBROUTINE
930 INPUT"ENR=" ;NR:NR=NR/10
940 TE=TO*(10^NR+1)
950 INPUT"NOISE BANDWIDTH IN MHZ=" ;BW:BW=BW*1000000!
960 PN=K*TE*BW
970 RETURN

```

The noise specification conversion program.

LOWEST PRICED, HIGHEST QUALITY ATTENUATORS - BNC \$11.00 1-9 EA., SMA \$14.40 10 EA. AND TERMINATIONS - BNC \$5.60 10 EA., SMA \$5.60 10 EA., MIL, HI-REL. NETWORKS

Model Number (2)	Impedance Ohms (Power W)	Frequency Range	BNC	TNC	N	SMA	UMF	PC
Fixed Attenuators 1 to 20 dB								
AT 50(3)	50 (1 SW)	DC 1-50Hz	14.00	20.00	20.00	18.00	—	—
AT 51	50 (1 SW)	DC 1-50Hz	11.00	18.00	15.00	14.00	—	12.00
AT 52	50 (1 SW)	DC 1-50Hz	14.50	20.50	20.50	19.50	—	—
AT 53	50 (1 SW)	DC 3-300Hz	14.00	17.00	—	15.00	—	—
AT 54	50 (1 SW)	DC 4-20Hz	—	—	—	18.00	—	—
AT 55	50 (1 SW)	DC 4-20Hz	—	—	—	14.40 (10 EA)	—	—
AT 75 or AT 90	75 or 93 (1 SW)	DC 1-50Hz (1750MHz)	11.50	20.00	20.00	18.00	—	—
Detector Mixer: Zero Bias Schottky								
CD-51 75	50 75	01-4 20Hz	54.00	—	—	54.00	—	—
DM-51	50	01-4 20Hz	—	—	—	64.00	—	—
Relative Impedance Transformers: Minimum Loss Pads								
RT 50 75	50 to 75	DC 1-50Hz	10.50	19.50	19.50	17.50	—	—
RT 50 93	50 to 93	DC 1-50Hz	13.00	19.50	19.50	17.50	—	—
Terminations								
CT 50 (3)	50 (1 SW)	DC 4-20Hz	11.50	18.00	15.00	17.80	—	—
CT 51	50 (1 SW)	DC 4-20Hz	9.50	12.00	10.50	9.50	—	—
CT 52	50 (1 SW)	DC 2-50Hz	10.50	15.00	15.00	13.00	15.50	—
CT 53/58	50 (1 SW)	DC 4-20Hz	9.50 (10 EA)	—	—	8.50 (10 EA)	—	—
CT 54	50 (2 SW)	DC 2-50Hz	14.00	15.00	15.00	17.50	—	—
CT 75	75 (1 SW)	DC 2-50Hz	10.50	18.00	15.00	13.00	18.50	—
CT 93	93 (1 SW)	DC 2-50Hz	13.00	18.00	—	15.00	18.50	—
Mismatched Terminations: 1:5 to 1:1, Open Circuit, Short Circuit								
MT 51	50	DC 1-50Hz	45.50	45.50	45.50	45.50	—	—
MT 75	75	DC 1-50Hz	—	—	—	—	—	—
Feed thru Terminations: Shunt resistor								
FT 50	50	DC 1-50Hz	10.50	19.50	19.50	17.50	—	—
FT 75	75	DC 500MHz	10.50	19.50	19.50	17.50	—	—
FT 90	93	DC 1-50MHz	13.00	19.50	19.50	17.50	—	—
Directional Coupler: 30 dB								
DC 500	50	250-500MHz	60.00	—	64.00	—	—	—
Resistive Decoupler: series resistor or Capacitive Coupler: series capacitor								
RD or CC 1000	1000 (1000PF)	DC 1-50Hz	12.00	18.00	18.00	17.00	—	—
Adapters								
CA 50 (N to SMA)	50	DC 4-20Hz	13.00	13.00	13.00	13.00	—	—
Inductive Decouplers: series inductor								
LD 15	0.17uH	DC 500MHz	12.00	18.00	18.00	17.00	—	—
LD 68	6.8uH	DC 55MHz	12.00	18.00	18.00	17.00	—	—
Fixed Attenuator Sets: 3, 6, 10, and 20 dB in plastic case								
AT 50-82T (3)	50	DC 1-50Hz	60.00	84.00	84.00	78.00	—	—
AT 51-82T	50	DC 1-50Hz	48.00	84.00	84.00	80.00	—	—
Reactive Multicouplers: 2 and 4 output ports								
TC 125-2	50	1.5-125MHz	64.00	—	67.00	67.00	—	—
TC 125-4	50	1.5-125MHz	67.00	—	81.50	81.50	—	—
Resistive Power Dividers: 3, 4 and 9 ports								
RC 3 50	50	DC 2-50Hz	84.00	84.00	—	84.00	—	—
RC 4 50	50	DC 500MHz	84.00	84.00	—	84.00	—	—
RC 8 50	50	DC 500MHz	84.00	84.00	—	84.00	—	—
RC 3-75 475	75	DC 500MHz	84.00	84.00	—	84.00	—	—
Double Balanced Mixers								
DBM 1000	50	5-1000MHz	61.00	—	71.00	81.00	—	34.00
DBM 500PC	50	2-500MHz	—	—	—	—	—	34.00
RF Fuse: 1:8 Amp and 1:16 Amp								
FL 50	50	DC 1-50Hz	12.00	18.00	45.50	17.00	—	—
FL 75	75	DC 1-50Hz	—	—	—	17.00	—	—

NOTE 1) Critical parameters fully tested and guaranteed. Fabricated from Mil Spec. High-Rel. resistors. Schottky diodes. Mil Spec. plated parts and connectors in nickel, silver, and gold. 2) See catalog for complete Model Number. Specify connector boxes. Specials available. 3) Calibration marked on label of unit. 4) Price subject to change. 1988B without notice. Shipping \$5.00 Domestic or \$25.00 Foreign on Prepaid Orders. Delivery is stock to 30 days ARO.

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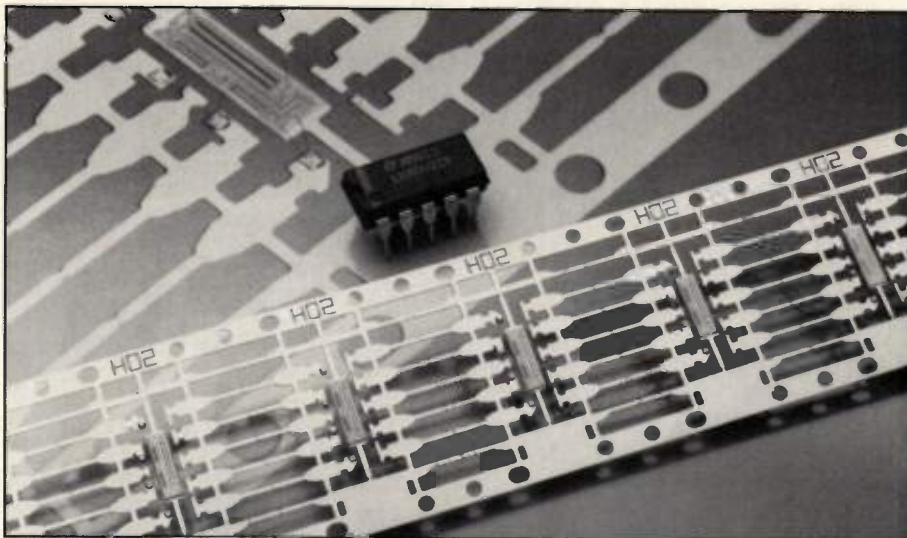


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The LH4002 is a high speed unity gain buffer designed to provide high current drive capability at frequencies from DC to over 200 MHz with a bandwidth of at least 100 MHz. It operates on a ± 5 V supply and features a minimum slew rate of 1000 V/ μ s with the ability to drive 50 and 75 ohms directly. Additional AC specifications include a phase linearity of 2 degrees (at a bandwidth from 1 to 20 MHz) and distortion of less than 0.1 percent.

The LH4002CN is available in a 10-pin plastic DIP while the LH4002CH, LH4002H and the LH4002H-MIL are encased in 8-pin metal TO-5 packages. The 10-pin plastic DIP uses lead frame construction which provides good power dissipation to ambient.

This buffer is intended for a range of high speed applications such as video distribution, impedance transformation, and load isolation. In current booster applications within an op-amp loop, the device can increase the output current



capability of existing op-amps.

In quantities of 100, the LH4002CN and LH4002H cost \$9.50, the LH4002H (Military Temp) is \$19.25 and the LH4002H-

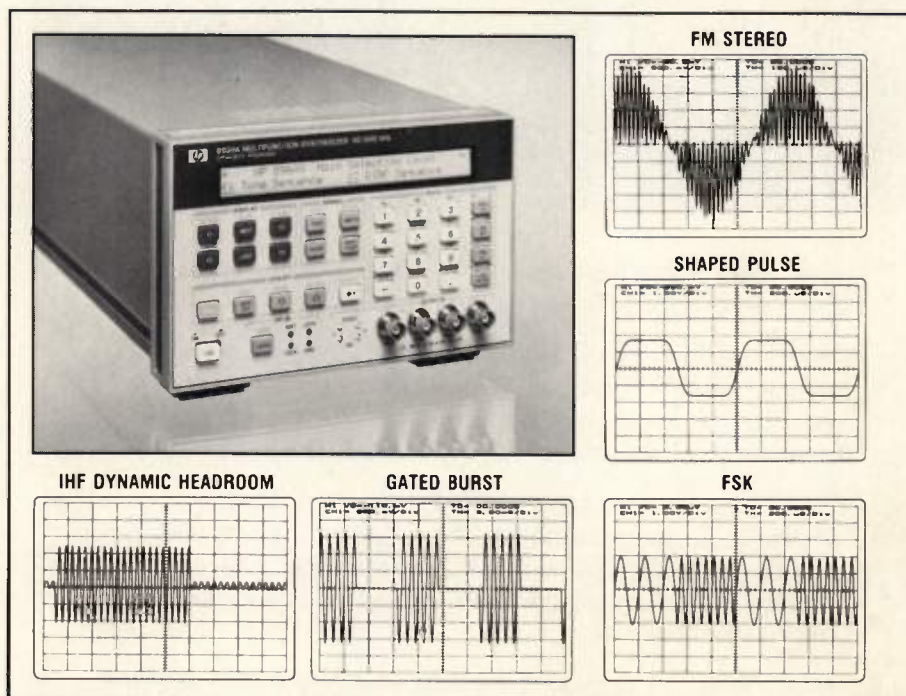
MIL (Military Processed) is priced at \$46.40. **National Semiconductor Corporation, Santa Clara, CA. Please circle INFO/CARD #220.**

HP Introduces A Multifunction Synthesizer

The HP8904A multifunction synthesizer is an instrument that digitally creates complex signals from six basic waveforms. It can be used as a function generator, modulation source, or a stimulus for audio circuits. Signals for VOR, ILS, FM stereo or communications signalling can also be created. The instrument starts as a synthesizer that digitally generates sinewaves to 600 kHz, square, ramp, and triangle signals to 50 kHz, and white noise and DC signals. The standard 8904A offers the precision of a digital synthesizer with 0.1 Hz resolution, an electronic 50 ohm output and standard HP-IB.

Three option packages are available for this device. Option 001 adds three synthesizers or channels which can either modulate the first synthesizer or be summed to the output. Option 002 adds a second output to the instrument. For digital modulation and fast hop applications, Option 003 adds external timing control.

Complex signals can be created with Option 001 by adding synthesizers which can either AM, FM, PM, DSBSC or pulse modulate the first synthesizer. Or, any can be summed with the first channel at the output. All four channels can be set to generate different waveforms, frequencies, amplitudes, and phase offsets simultaneously. Option 002 adds a sec-



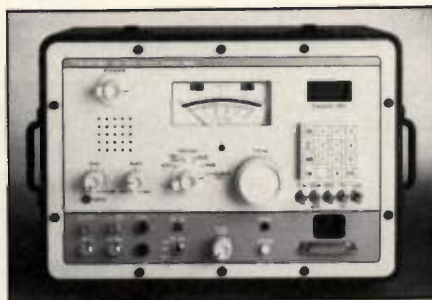
ond identical synthesizer and floating output section. With phase accuracy between the two outputs specified to be better than ± 0.10 degrees, the instrument can test phase detectors, servo systems, shaft encoders, sonar, and other two port

devices.

The HP8904A is priced at \$2600 while Option 001 costs \$1500, Option 002 costs \$1200 and Option 003 is approximately \$500. **Hewlett-Packard Company, Palo Alto, CA. INFO/CARD #219.**

4W Operational Amplifier

The WA01 features a power bandwidth of 150 MHz and a slew rate of 4000V/us. It has a maximum voltage offset of 10 mV, drift of 50 uV/C and non inverting bias current of 20 uA. The internal 1.5 k ohm feedback resistor provides a transimpedance function of 1.5 V/mA in respect to the inverting input. The WA01 is \$160.85 in single quantities. **Apex Microtechnology Corp., Tucson, AZ. INFO/CARD #216.**



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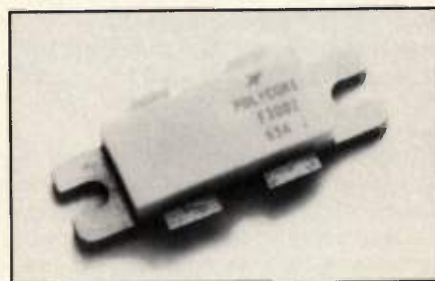
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Gold Metalized FET

Polycore RF Devices introduces a series of gold metalized FETs at 300 W. The F3000 Series uses layered titanium-tungsten-gold. It offers push-pull linear or single ended high power configurations. The devices operate from 1 MHz to 300 MHz. **Polycore RF Devices, Newbury Park, CA. INFO/CARD #217.**

Scalar Network Analyzer

The Wiltron 561 makes measurements of transmission, return loss, and power over the 10 MHz to 40 GHz range without using external controllers. The instrument displays at one time any two of four input channels over a specified 71 dB dynamic range. Noise floor is less than -62 dBm providing a greater than 76 dB range.



Data interpretation at low signal levels is enhanced with smoothing and averaging. **Wiltron Company, Morgan Hill, CA. INFO/CARD #215.**

Flexible Cables

Four types of cables namely FE 10, FE 12, FE 19 and FE 25 are being introduced by Adams Russell's Antenna and Microwave Division. The FE 10 is rated from 30 MHz to 74 GHz with an insertion loss of 1.3 dB and VSWR of 1.25:1. FE 12 goes from 30 MHz to 55 MHz and has an insertion loss of 0.9 dB. The frequency range for FE 19 is 30 MHz to 30 GHz with an insertion loss of 0.48 dB. 30 MHz to 23.5 GHz is the range for Model FE 25. Its insertion loss is 0.42 dB. Note that the frequency ranges are dependent on connector style. **Adams Russell, Antenna and Microwave Division, Amesbury, MA. INFO/CARD #214.**

Miniature Linearized Attenuator

The MA2694 miniature voltage variable attenuator incorporates a hybridized linearizer which reduces the package size. They are offered with 40 dB or 60 dB dynamic range in standard or sub-octave bands from 2 to 18 GHz. **M/A-COM Solid State Division, Hudson, NH. INFO/CARD #213.**

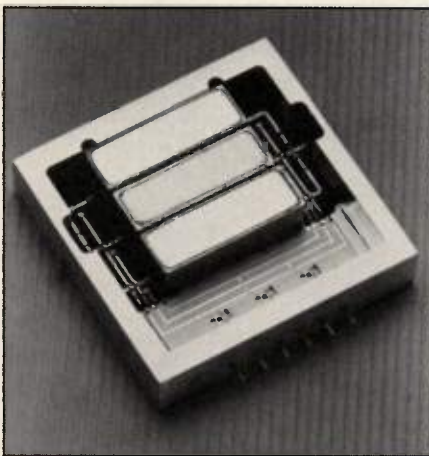
Fiber Optic Chip Set

The chip set provides solutions for encode and decode in high speed digital fiber optic communications systems. The SP9960 encoder /LED driver accepts either TTL or ECL level signals and encodes the clock and data signals for transmission. The SL9901 transimpedance amplifier converts the output from a PIN diode detector into a voltage suitable for driving the amplifier input stage of the decoder circuit. SP9921 is a decoder which operates from 20 MB/sec to 50 MB/sec. It recovers the clock and data from the amplified input signal and provides ECL-level outputs. The quantity prices are \$21.04 each for the SP9960, \$7.78 each for the SL9901 and \$38.69 for

the SP9921. **Plessey Semiconductors, Irvine, CA. INFO/CARD #211.**

Switched Filter Banks

Comstron has developed switched filter banks for use in systems where contiguous, overlapping, variable bandwidth



or linear phase (constant delay) designs are required. The frequency range is 1 MHz to 12 GHz. **Comstron Corporation, Melville, NY. INFO/CARD #212.**

Linear Power Amplifier

Model A2001 class A linear power amplifier operates from 10 to 2000 MHz. It provides a minimum of 1 W output with a gain flatness of ± 1.5 dB. **Amplitech, Paramus, NJ. INFO/CARD #209.**

RF Tuning Network

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- 80dB of gain control range
- SIL package

TDA1011

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- Excellent noise rejection
- SIL package
- Wide supply voltage range (3.6V-20V)

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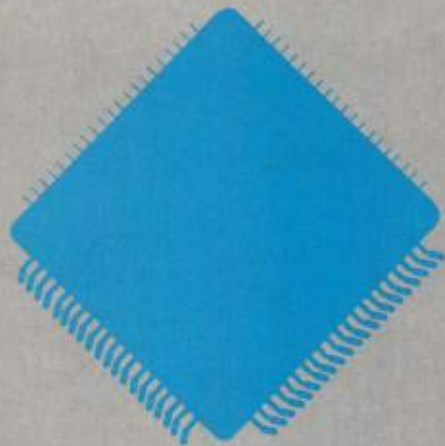
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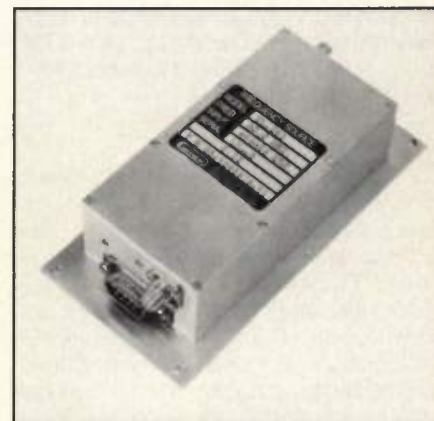
rf products *Continued*



analog and TTL computer interface, RF generator control, and DC bias controllability. The digital system is fully automatic with manual overrides. **Mercator Control Systems, Inc., Santa Ana, CA. INFO/CARD #208.**

PLL Crystal Oscillator

Greenray Industries introduces a phase locked loop crystal oscillator for frequencies from 10 to 100 MHz. The spur-



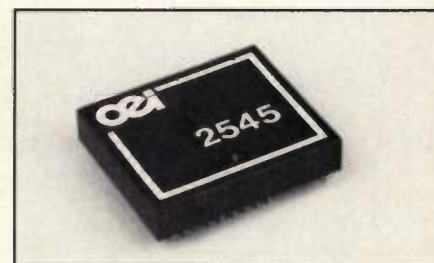
ious specification is -70 dBc at $F_c \pm 5$ MHz. **Greenray Industries, Inc., Mechanicsburg, PA. INFO/CARD #210.**

Octave Band VCOs

Model V1 voltage controlled oscillators provide full octave tuning in bands to 200 MHz. These low voltage units feature output buffering with output power at 13 dBm or ECL levels. **Luff Research, Jackson Heights, NY. INFO/CARD #207.**

Video Enhancement Module

The 2545 enhancement module is a

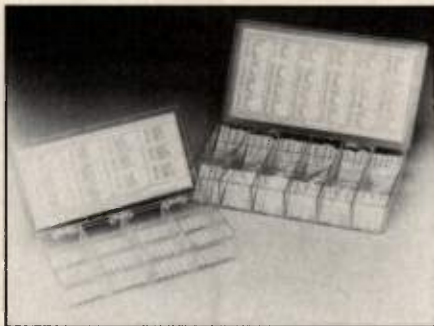


September 1987

video signal processing block that contains a high speed logarithmic amplifier and support circuitry for interfacing a logarithmic expansion or compression function into a video system. For 1 to 9 pieces the 2545 is priced at \$637. **Optical Electronics, Inc., Tucson, AZ.** Please circle INFO/CARD #206.

Chip Resistor and Capacitor Kits

Communications Specialists introduces a chip resistor (CR-1) and chip capacitor (CC-1) kit. The CR-1 contains 1540 pieces composed of 10 chip resistors for every 5 percent value from 10 ohms to 10 M ohms plus 8 values of 0, 10, 100, 1 k, 10



k, 1 M, and 10 M ohms. CC-1 contains 365 pieces composed of 5 chip capacitors for every 10 percent value from 1 pF to .33 uF plus 5 additional capacitors for 1 pF, 10 pF, 100 pF, 1000 pF, .01 uF and .1 uF. Each kit sells for \$49.95. **Communications Specialists, Inc., Orange, CA.** INFO/CARD #205.

High Voltage Crosspoint Switches

SD1500BD is a high speed 600 V enhancement-mode vertical D-MOS FET in a TO-92 package for low current high voltage switching applications. It is also available in a die form. For 100 pieces and up, individual units cost \$0.95. **Topaz Semiconductor, San Jose, CA.** Please circle INFO/CARD #204.

Miniature Voice Scrambler

The VPU-7 features a wide input dynamic range, an antialiasing input filter, and a 6 pole output filter. It has a crystal controlled single inversion, low distortion audio recovery and an optional flip-flop with board extension for momentary push



button control. The dimensions are 1.1" x .69" x .18". **Midian Electronics, Inc., Tucson, AZ.** INFO/CARD #203.

Tempest Power Line Filter

The Standex TA and TB Series of power line filters features extended insertion loss to 40 dB/.05 MHz, 60 dB/.1 MHz, 70 dB/.15 MHz, and 80 dB/.3 MHz. Power entry module provides two pole On and OFF switch, fusing, power cord connector, (shielded or unshielded) and flush mounting with optional R.F. gasket. **Standex Electronics, Riviera Beach, FL.** Please circle INFO/CARD #202.

RF Power Amplifier

The Model 4200 delivers over 1000 W of CW power and is protected against excessive VSWR. The gain is 60 dB minimum with a gain flatness of ± 1 dB. This Class AB unit has a frequency range of 400 to 450 MHz. **American Microwave Technology, Inc., Fullerton, CA.** Please circle INFO/CARD #201.

ECL VCXO

Piezo Technology introduces a hybrid ECL voltage controlled crystal oscillator. The CHX1004 operates at a frequency of 44.736 MHz. It features a 10 K ECL output signal with a linearity better than ± 10 percent, maximum rise and fall time of 2 ns, and a frequency stability of ± 25 ppm over the -40°C to $+70^\circ\text{C}$. The dimensions for the device are .0803" x .0502" x .0225". **Piezo Technology, Inc., Orlando, FL.** INFO/CARD #200.

Amplifiers With Blindmate Connectors

Aydin Vector introduces the AMX Series of custom modular amplifiers with blindmate connectors. The frequency range is



from 5 to 1500 MHz, noise figure range is 2.5 to 10 dB, gain range is from 10 to 55 dB and the power output range is -2 to $+22$ dBm. The device is available in TO-8, TO-12, and 4 pin dip with miniature

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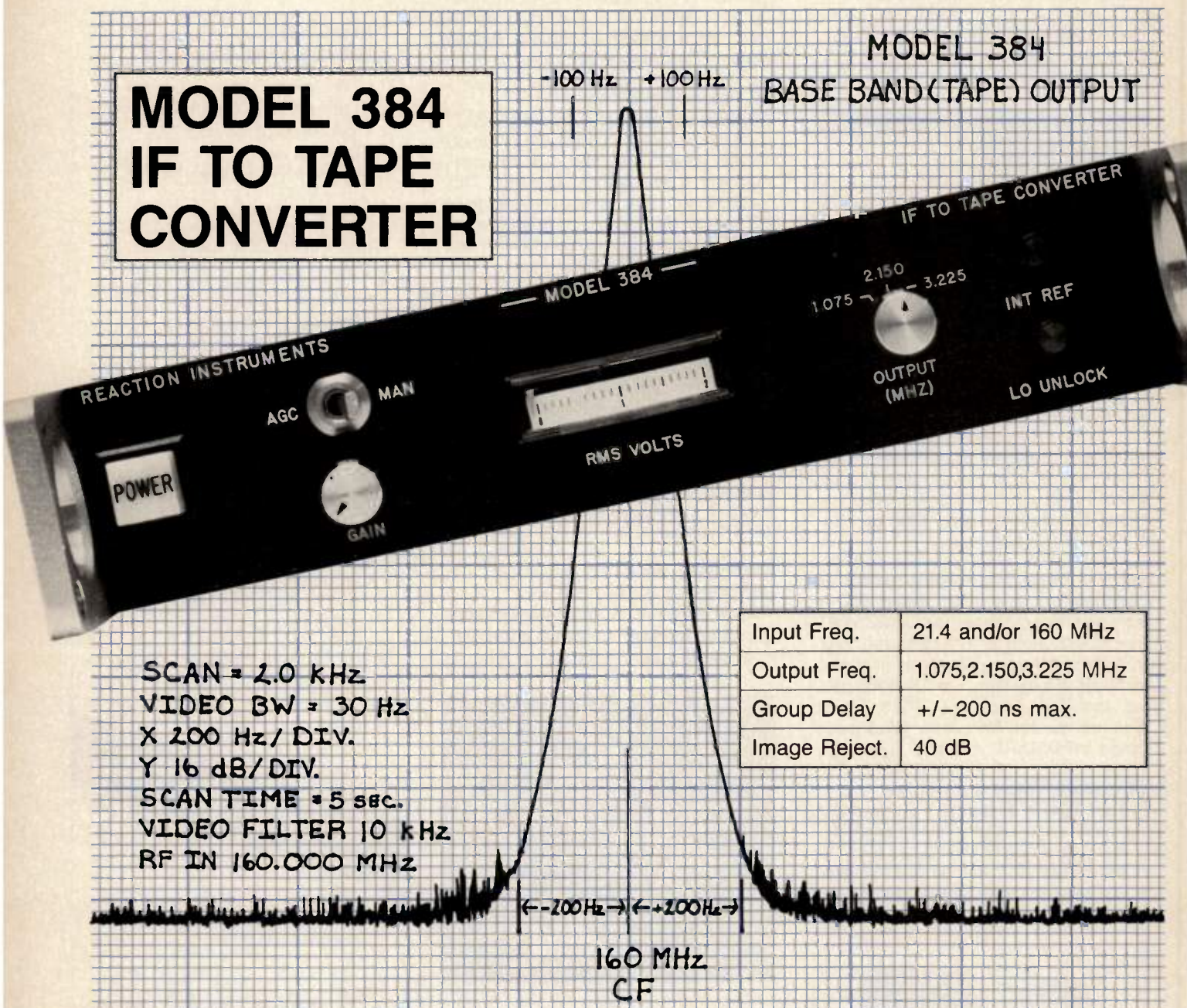
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INFO/CARD 58

WRH

SMA and blindmate connected housing. Aydin Vector Division, Newton, PA. INFO/CARD #199.

Digital Storage Oscilloscope

Rapid Systems introduces the R2000 2-channel 20 MHz digital storage oscilloscope that uses the PC for display and storage. The features include 2 channel simultaneous acquisition, 20 MHz sam-



pling rate per channel, switch selectable 1 M ohm or 50 ohm BNC input, programmable gain and sample rate, menu driven operation, and screen dumps to the printer. The system costs \$3495 and is tailored for IBM and compatible computers. Rapid Systems, Inc., Seattle, WA. INFO/CARD #198.

Low Noise GaAs Amplifier

Tachonics Corporation introduces the TCWL-0100 low noise GaAs monolithic IC that operates over the .05 to 5 GHz frequency range. Within the 0.05 to 4 GHz range, it offers a 12 dB gain and a typical N.F. of 3.5 dB. The output power at 1 dB compression power is 14 dBm. It is available in chip, stripline or surface mount packages. The packaged units also come in single or dual supply configurations. Prices in 100-piece quantities begin at \$17.80. Tachonics Corporation, Plainsboro, NJ. INFO/CARD #197.

Digital Quadrature Detector

Mirage Systems introduces the Model 850A digital quadrature detector. It operates from 50 to 200 MHz with a



dynamic range of greater than 80 dB. The resolution is 16 bits for amplitude and phase. Mirage Systems, Sunnyvale, CA. INFO/CARD #196.

Transimpedance Operational Amplifier

Analog Devices introduces a transimpedance operational amplifier that uses current rather than voltage feedback so that the bandwidth is relatively independent of gain. The hybrid AD9610 offers a 100 MHz (-3dB) unity gain bandwidth, 95 MHz bandwidth at a gain of -10, and 75 MHz at a gain of -20. The slew rate is 3500 V/us. Over a frequency range of 5 MHz to 150 MHz noise is measured at 0.7 nV/√Hz for voltage and 23 pA/√Hz for current. Analog Devices, Norwood, MA. INFO/CARD #195.

Bias Tee

Alford Manufacturing introduces a bias tee, Model 9460-1.0- xxxx, which covers from 500 MHz to 1.5 GHz. The RF loss is measured at 0.2 dB. Alford Manufacturing Company, Woburn, MA. Please circle INFO/CARD #194. ☐

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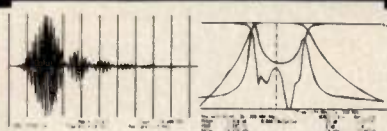
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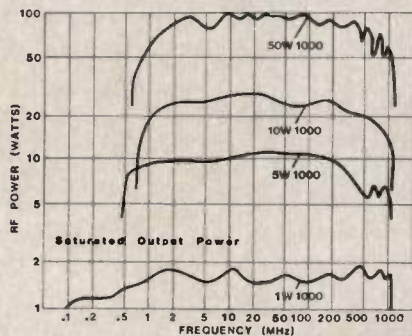
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INFO/CARD 61

Amplifiers and Signal Processing Components

Cougar Components has published its 1987 catalog entitled *RF and Microwave Cascadable Amplifiers and Signal Processing Components 0.3 to 2400 MHz*. Included are complete specifications and graphs for their line of amplifiers and voltage controlled attenuators. All assembly techniques, parts, and materials are designed to meet the screening and manufacturing requirements of MIL-STD-883. **Cougar Components, Sunnyvale, CA. INFO/CARD #179.**

Solving Coil Design Problems

Magnecraft has developed a graphical method for solving coil design problems. New formulae compare wire gauge to coil resistance, coil turns, power dissipation and ampere turns. **Magnecraft Electric Company, Northbrook, IL. Please circle INFO/CARD #178.**

Signal Processor Information

A set of technical information on the TS68930 Digital Signal Processor is available from Thomson-Mostek. The literature set comprises of a data sheet, a DSP comparison of the TMS 32020 vs TS68930, a TS68930 DSP module library providing a list of software modules and macros, and a real time algorithms application note. **Thomson-Mostek, Carrollton, TX. INFO/CARD #177.**

Pulse Generators and Accessories

Avtech Electrosystems introduces its 1987 general catalog of nanosecond waveform generators and accessories. It describes over 150 models of ultra high speed pulse generators, impulse generators, monocyte generators, power splitters, and scope probes covering the PRF range of 0 to 250 MHz, rise times from 40 ps to 10 ns, pulse widths from 130 ps to 100 us and amplitudes from 2 to 500 V. **Avtech Electrosystems Ltd., Ottawa, Ontario, Canada. INFO/CARD #176.**

MIC Microwave and IF/RF Catalog

RHG Electronics Laboratory has released a catalog (No.400) containing product and technical information on their complete line of MIC Microwave and IF/RF products. It incorporates advancements in component, circuit and packaging technology for IF/RF products, MIC mixers and mixer preamps, microwave components, and microwave links and monopulse receivers. A technical information and article section is also included with information on log amps, phase and gain matching, signal-to-noise ratio, fine gain linearity measurements

and monopulse receivers. Test methods for measurement of standard specifications are included. **RHG Electronics Laboratory, Inc., Deer Park, NY. Please circle INFO/CARD #174.**

RF Filters Catalog

This catalog contains information on powerline shielded room filters, RFI/EMC filters for various current levels and ranges, filter discharge units, line impedance networks, and secure communications and signal line filters. RF filtered and shielded circuit breaker panel boards and power factor correction coils are included. **Filtron Company, Inc., Old Bethpage, NY. INFO/CARD #173.**

DPAK Literature Package

Motorola introduces the DPAK literature package which includes literature on all discrete devices currently available in surface mount packaging (DPAK). Bipolar power transistors, TMOS power MOSFETs, and Schottky and ultra-fast rectifiers are featured. The package includes a DPAK update, a surface mount selector guide, and article entitled *DPAK, The Power Package For Surface Mount*, a tape and reel data sheet for surface mount devices, and DPAK product data sheets including 26 bipolar power transistors, 8 power MOSFETs and 16 rectifiers. **Motorola, Inc., Phoenix, AZ. Please circle INFO/CARD #172.**

Test and Measurement Instrumentation Catalog

This 16 page catalog outlines performance features, applications, and specifications for 22 instruments. Products include digital multimeters, and communications test sets. **Solartron Instruments, Elmsford, NY. INFO/CARD #171.**

Short Form Catalog on Instruments

A short catalog covering its line of Test and Measuring instruments has been issued by Iwatsu Instruments. It presents features and specifications on oscilloscopes, function generators, scope wagons, trace recording camera systems, a selection chart of oscilloscope accessories, and a passive and active probes chart. **Iwatsu Instruments, Inc., Carlstadt, NJ. INFO/CARD #170.**

Filter Catalog

A line of active and passive signal-conditioning filters is presented in this catalog from TTE. Filter designs offered are Butterworth, Chebyshev, Gaussian, Bessel, anti-aliasing, programmable and notch types, plus TTE's own designs. The frequency range covered is from 0.1 Hz

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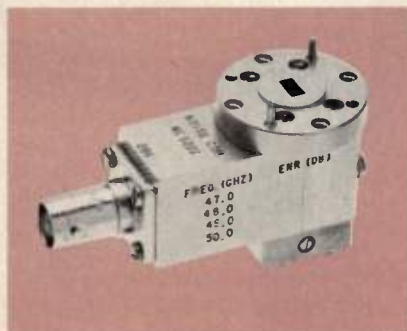
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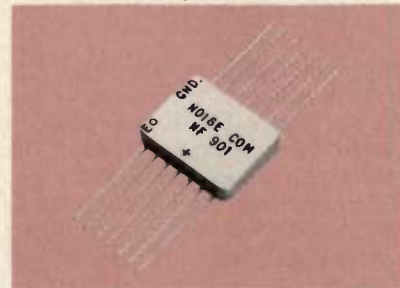
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to 500 MHz. Documentation provided includes specifications, attenuation curves, response curves, response comparisons, case drawings and mounting dimensions. **TTE, Inc., Los Angeles, CA. Please circle INFO/CARD #169.**

Scalar Network Analyzer Brochure

The brochure describes a scalar network analyzer that measures transmission, return loss, and power over the 10 MHz to 40 GHz range. Also contained are typical measurement displays, a test setup diagram, front panel functions, applications, accuracy tables, and a list of available models. **Wiltron Company, Morgan Hill, CA. INFO/CARD #168.**

Conductive Silicon Rubber Brochure

Performance data of MOX-C/S conductive silicon rubber for EMI/RFI shielding applications is described in a series of brochures available from Moxness Products. These shielding products are made from electrically conductive rubber. **Moxness Products, Inc., Racine, WI. Please circle INFO/CARD #167.**

Coils and Inductive Components Catalog

The catalog lists primary operational and physical parameters of Delevan's molded, shielded, and variable RF coils with inductance values from 0.022 μ H to 150 mH. It also lists toroidal inductors, power chokes, and micro-i chip inductors. **Delevan Division, American Precision Industries, East Aurora, NY. Please circle INFO/CARD #166.**

IEEE 488 Catalog

IOtech has published its 1987 catalog of IEEE 488 interfaces. The products included are IOtech's Personal 488/2 interface, the expanded 488 bus expander, the Extender 488/F fiber optic bus extender, the Chrono 488 bus clock, and the COM488, an IEEE printer/plotter interface. **IOtech, Inc., Cleveland, OH. Please circle INFO/CARD #165.**

Chip Capacitor Data Sheet

A MIL-C-55681 and MIL-C-11272 multi-layer chip capacitor brochure #62-08 is available from Murata Erie North America. Included is information on CDR01-06, CRD11-14, CY81-84 and the CDR21 Series with M failure rate level approval. Also listed are performance specifications with respect to capacitance, insulation resistance, voltage temperature limits, life test and resistance to soldering heat. **Murata Erie North America, Inc., State College, PA. INFO/CARD #175.**

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ANDERSEN: We build SAW hybrid oscillators to phase noise specs of -95dBc/Hz at 100 Hz offset from the carrier to less than -160dBc/Hz at 100KHz.

QUESTION: Are there high NRE costs associated with SAW oscillators?

ANDERSEN: Not with us. We've developed designs for fixed frequency and VCO applications which can cover the entire 100 MHz to 1 GHz range. So we normally don't charge any NRE. That's why we're the leaders in SAW oscillators.

QUESTION: Where are SAW oscillators being used?

ANDERSEN: Our fixed-frequency SAW oscillators are being used in IFF and air traffic control systems, as local oscillators for frequency up/down converters in radar and communications systems, as master clocks for digital communications systems, just to name a few of the applications.

Our SAW VCO's are used in a wide range of Phase Locked Loops including timing recovery circuits for fiber optic data transmission systems where the oscillator reestablishes the timing of the data stream at the end of the fiber optic lines.

QUESTION: How do I get started?

ANDERSEN: Just give us a call. We'll do everything we can to meet your needs. If you simply want more information, send for our handbook on SAW Hybrid Oscillators. Write to Andersen Laboratories, 1280 Blue Hills Avenue, Bloomfield, CT 06002, Phone (203) 242-0761/ TWX 710-4253 2390

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rfdesign

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1988 Directory Issue

1988 directory

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9



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Interface: BCD par. or GPIB
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Price: \$4,800.00*

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Switching: 5-20 μ s
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INFO/CARD 3

This is the *Quick and Easy* Directory



By Gary A. Breed
Editor

The 1988 *RF Design* Directory issue is ready to serve you. In this magazine are product listings for everything in RF from amplifiers to Z-bridges. The simple arrangement of product categories and companies should make it easy to put together a list of firms to contact when you need a particular item.

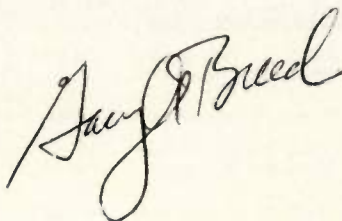
The products are listed by major categories (components, subassemblies, test equipment, etc.), which are then divided into logical subgroups (small-signal FETs, spectrum analyzers, shielding materials, etc.). These breakdowns have been carefully designed to provide necessary information without forcing you to dig through the dozens of nearly-identical categories found in other product directories.

Companies are listed alphabetically in a single, uninterrupted section. This section is marked with a black band on the outside edge, making it impossible to miss. Finding the address, telephone and contact information on the companies you want is a cinch.

A special feature in this year's Directory is a complete listing of every article published in *RF Design's* ten-year history! *Keep this issue, since the complete listing may not be repeated.* Starting in 1989, our December issue will contain a bibliography of articles for the previous two years.

We continue the Design Guide section introduced last year with some concise notes on coupled-resonator bandpass filters, plus a useful method to keep correct units in propagation path loss formulae.

Finally, there is a Comment Card in the back — this is for you to tell us how you use the Directory, and to offer suggestions for making it more valuable to you and your fellow RF engineers. We hope to hear from you!



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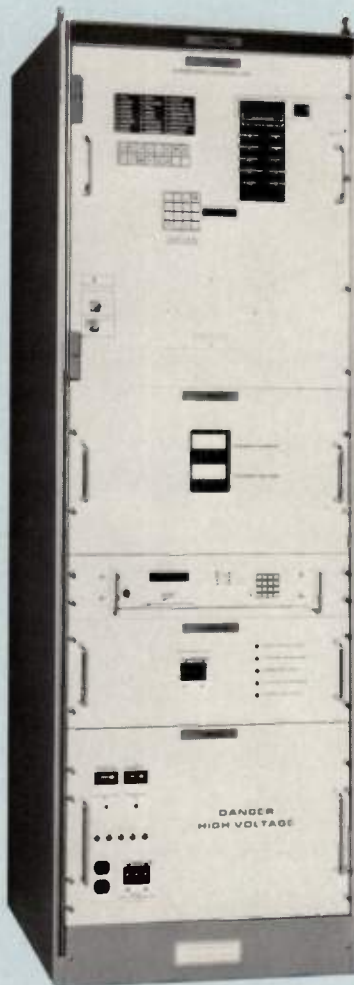
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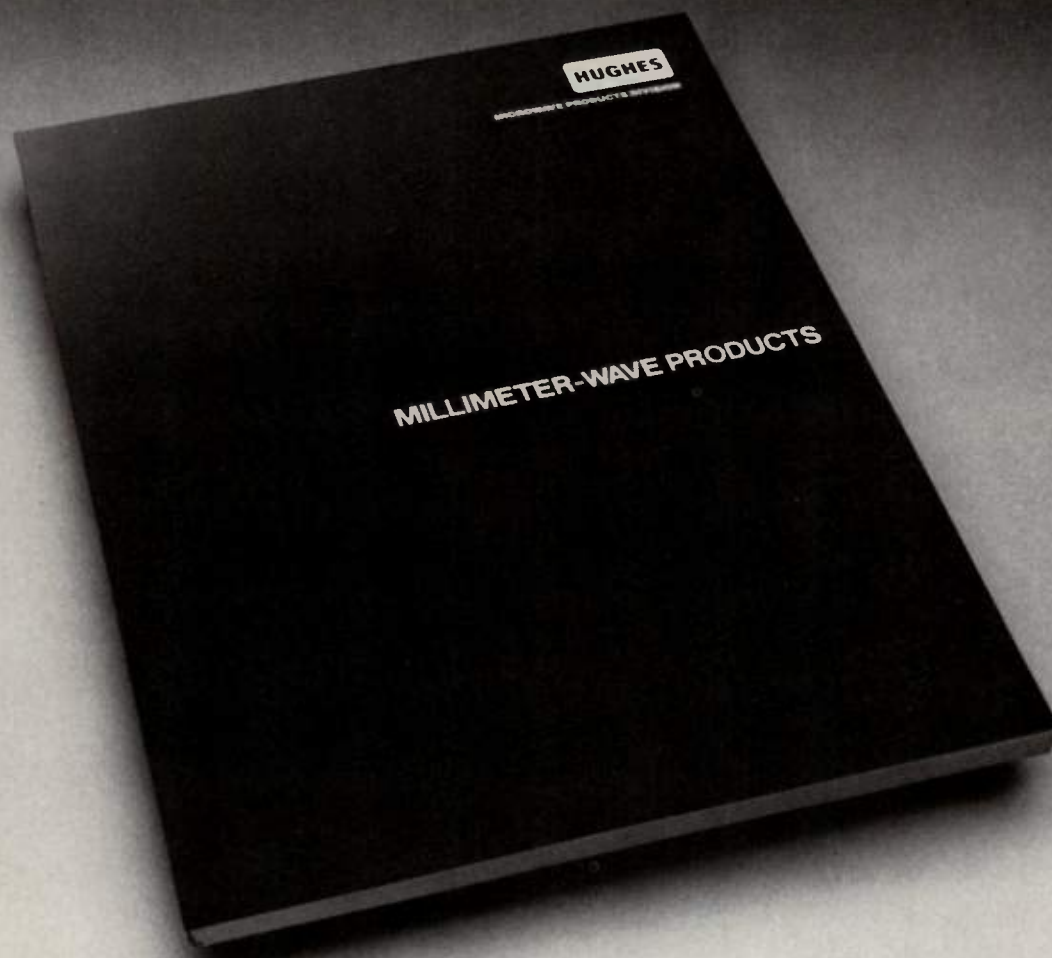


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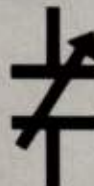
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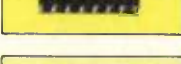



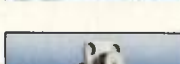
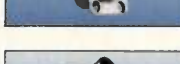






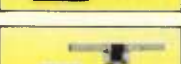




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TERMINATIONS    	Coaxial <ul style="list-style-type: none"> • 2-500 watts • DC-26.5GHz • Ideal for calibration and debugging Flange Mount <ul style="list-style-type: none"> • 2-500 watts • External mounting replicates connectors • Compact, easy to assemble Stripline <ul style="list-style-type: none"> • 1-3 watts • Pill design mounts internally • Small size, simple installation Surface Mount <ul style="list-style-type: none"> • Up to 150 watts • DC-4GHz • For micro-miniature packages • Ideal for high frequency systems 	ATTENUATORS    	Switchable <ul style="list-style-type: none"> • 1-15 dB • Pre-soldered pads on Al₂O₃ • Surface mounting configurations • Ideal for fast accurate circuit adjustment Programmable <ul style="list-style-type: none"> • 750MHz-8GHz • Automatic compensation for signal strength • Various configurations available Thin Film Chip <ul style="list-style-type: none"> • 5dB to 20dB in 5dB steps • DC-12GHz • Unprecedented accuracy • Low cost, small size Hi Rel Chip <ul style="list-style-type: none"> • 1-20 dB • Full test documentation • Low cost, small size 	DC BLOCKS, SHORTS, CAPACITORS AND RESISTORS    	Flange Mount <ul style="list-style-type: none"> • 1-18GHz • 100V peak • Ideal for systems where coax blocks can't be used SMA <ul style="list-style-type: none"> • 1-18GHz • 100V peak • Low VSWR HIGH POWER RESISTORS <ul style="list-style-type: none"> • 25-250 ohms • DC-2GHz • Up to 150 watts ROD RESISTORS <ul style="list-style-type: none"> • 10-500 ohms • 125°C watts • AL₂O₃ or BeO substrates
ATTENUATORS     	SMA Coax <ul style="list-style-type: none"> • 0-30dB • DC-18GHz • 2-10 watts Flange Mount <ul style="list-style-type: none"> • 0-20dB • DC-18GHz • Interchangeable transition pins Stripline <ul style="list-style-type: none"> • 0-20dB • Up to 3 watts • Ideal for limited space Surface Mount Chips <ul style="list-style-type: none"> • 1-20dB • DC-18GHz • Many terminals, lead choices Hi Power <ul style="list-style-type: none"> • 125 watts • DC-2GHz • Up to 60dB 	DC BLOCKS, SHORTS AND CAPACITORS  	Miniature Chip <ul style="list-style-type: none"> • 1-20dB • DC-18GHz • Many sizes, shapes, power ratings • Stable, predictable performance Stripline <ul style="list-style-type: none"> • Pill and chip configurations for small packages • Direct mounting onto stripline circuits • Fast, simple assembly Coax <ul style="list-style-type: none"> • SMA connector configuration for flexibility • Replace standard connectors for cost space savings • Small size, light weight 	  	THICK OR THIN FILM RESISTORS <ul style="list-style-type: none"> • 10-500 ohms • 0.1-100 watts • 100-300V • Many different designs CHIP RESISTORS <ul style="list-style-type: none"> • 10-10K ohms • 1/20-1 watt • Chips or tabs • Many standard sizes, custom configurations PILL RESISTORS <ul style="list-style-type: none"> • DC-12GHz • 1/4-3 watts • 10-250 ohms

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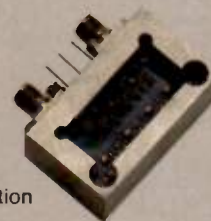


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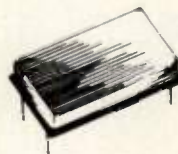
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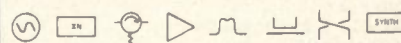
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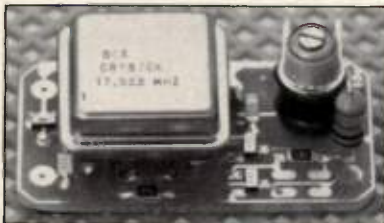
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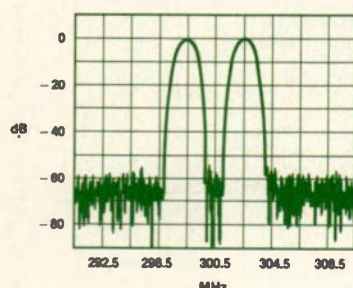
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VHF/UHF Filters

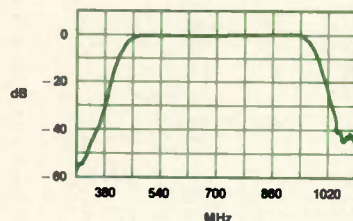
Cystal Technology offers production proven SAW filters, delay lines and resonators for:

Military Signal Processing



Phase/Delay Tracked Filters

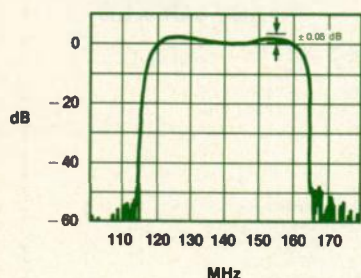
- Fundamental frequencies from 30 MHz to 1 GHz
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- Up to five filters/set. Phase tracked to $\pm 3^\circ$



Wide Band Delay Lines

- Center frequencies to 1 GHz
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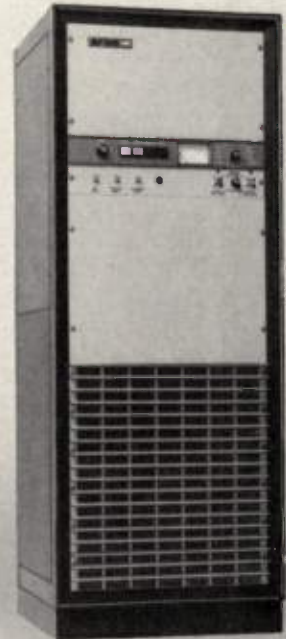
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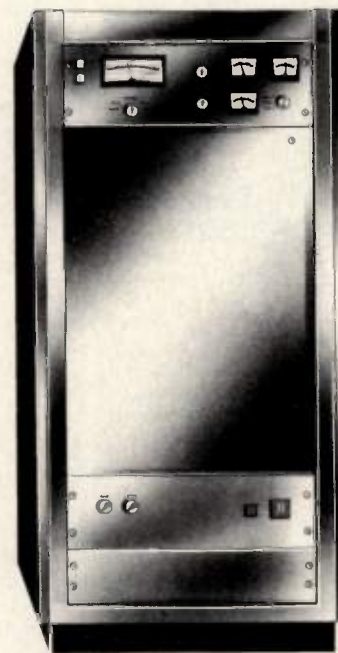
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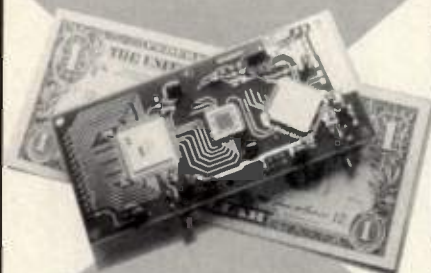
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ad on page 8**

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MITEQ, INC.
MM - WAVE TECHNOLOGY INC.
NEULINK
NURAD, INC.
PACIFIC MONOLITHICS
PACIFIC RESEARCH AND DEVELOPMENT
PASCALL ELECTRONICS LTD.
PICO PRODUCTS, INC.
PLESSEY SEMICONDUCTORS
PRECISION STANDARD TIME, INC.
REACTION INSTRUMENTS, INC.
REMTRON, INC.
RF MONOLITHICS INC.
RFD, INC.
RHG ELECTRONICS LABORATORY
ROSS ENGINEERING CORP.
SIGNAL ONE
SIGNATURE DESIGNS
SILTRONICS LTD.
SPREAD SPECTRUM SCIENCES
STEINBRECHER CORP.
SYSTRON DONNER CORP., MICROWAVE DIV.
TAUBER-DREYER CORP.
TECH COMM. INC.
TELE-TECH CORP.
TELEDYNE MICROELECTRONICS
TELEDYNE MICROWAVE
TELEMUS ELECTRONIC SYSTEMS INC.
TEST EQUIPMENT CORP.
TONY CHAPMAN ELECTRONICS LTD.
TRAK MICROWAVE CORP.
TRANS WORLD COMMUNICATIONS INC.
UHF ASSOCIATES
VARIAN RF SUBSYSTEMS DIV.
VITRO SERVICES CORP.
WATKINS-JOHNSON CO.
WEINSCHEL ENGINEERING
WILMANCO
XETRON CORP.
ZETA LABS

Transmission Components

RF MODEMS

APTRON LABORATORIES, INC.
CODEX
COMMUNITRONICS
DAJEM TECHNOLOGY CO.
EF JOHNSON
ELECTRO RENT CORPORATION
FG ENGINEERING CO.
JOSLYN DEFENSE SYSTEMS, INC.
LUNAR INDUSTRIES, INC.
MIDIAN ELECTRONICS
MITEQ, INC.
MOTOROLA SEMICONDUCTORS
REPCO INC.
RHG ELECTRONICS LABORATORY
SIGNATURE DESIGNS
SPREAD SPECTRUM SCIENCES
TEKTRON MICRO ELECTRONICS INC.
TELE-TECH CORP.
TELEMETRY TECHNOLOGIES, INC.
TEST EQUIPMENT CORP.
TRANS WORLD COMMUNICATIONS INC.
ZETA LABS

TONE ENCODERS/ DECODERS

A & A ENGINEERING
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ADVANCED COMMUNICATIONS SYSTEMS, INC.
APPLIED RESEARCH PRODUCTS OF SERVO CORP OF AMERICA
AYDIN VECTOR DIV.
DAJEM TECHNOLOGY CO.
ELECTRO RENT CORPORATION
EON INSTRUMENTATION, INC.
FG ENGINEERING CO.
HDS, INC.
JOSLYN DEFENSE SYSTEMS, INC.
NORCOMM CORP.
PALOMAR ENGINEERS
PICO PRODUCTS, INC.
PLESSEY SEMICONDUCTORS
REMTRON, INC.
SELECTONE CORP.
SIGNATURE DESIGNS
TEXAS SPECTRUM ELECTRONICS INC.

TRANSMITTERS

A & A ENGINEERING
A.R.F. PRODUCTS, INC.
AACOM DIVISION OF DATUM, INC.
ADAMS RUSSELL ELECTRONICS - ANZAC DIV.
AMERICAN LASER SYSTEMS, INC.
ANGHEL LABORATORIES, INC.
APPLIED RESEARCH PRODUCTS OF SERVO CORP OF AMERICA
APTRON LABORATORIES, INC.
AYDIN VECTOR DIV.
BRIMROSE CORP.
CARDION ELECTRONICS
COMARK COMMUNICATIONS, INC.
COMMUNITRONICS
DAJEM TECHNOLOGY CO.
EF JOHNSON
EHRHORN TECHNOLOGICAL OPERATIONS, INC.
ELECTRO RENT CORPORATION
EPSILON LAMBDA ELECTRONICS CORP.
ERBTEC ENGINEERING INC.
FG ENGINEERING CO.
FREQUENCY SOURCES, INC.
GENERAL SERVICES ENGINEERING, INC.
GILTRON INC.
HDS, INC.
HUGHES AIRCRAFT CO. MICROWAVE PRODUCTS DIV.
INSTRUMENTS FOR INDUSTRY, INC.
INTECH INC.
INTERAD LTD.
INTERNATIONAL MICROWAVE CORP.
JOSLYN DEFENSE SYSTEMS, INC.
KAPTRON INC.
L & M ENGINEERING SOUTH
LASER DIODE INC.
LUNAR INDUSTRIES, INC.
M/A-COM COMPONENTS MARKETING, INC.
M/A-COM MPD, INC.

MCL INC.
MERET INC.
MICROWAVE MODULES & DEVICES
MICROWAVE SEMICONDUCTOR CORP.
MICROWAVE SYSTEMS INC.
MM - WAVE TECHNOLOGY INC.
NEULINK
NURAD, INC.
ORIEL CORP.
ORTEL CORP.
REMTRON, INC.
REPCO INC.
RF MONOLITHICS INC.
RF PLASMA PRODUCTS
RFD, INC.
ROSS ENGINEERING CORP.
SIGNAL ONE
SIGNATURE DESIGNS
SPREAD SPECTRUM SCIENCES
STEINBRECHER CORP.
SYSTRON DONNER CORP., MICROWAVE DIV.
TEKTRON MICRO ELECTRONICS INC.
TELE-TECH CORP.
TEST EQUIPMENT CORP.
TRANS WORLD COMMUNICATIONS INC.
UHF ASSOCIATES
VENTRONICS
VITRO SERVICES CORP.
XETRON CORP.

TRANSMISSION COMPONENTS ANTENNAS

HF

AEL DEFENSE CORP.
AMPLIFIER RESEARCH
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BARKER & WILLIAMSON
CENTURION INTL INC.
CHILDS CORP.
CUSHCRAFT CORP.
DAJEM TECHNOLOGY CO.
DAYTON-GRANGER, INC.
DYMA ENGINEERING
FLAM & RUSSELL, INC.
JAMES MILLEN ELECTRONICS
JASCO INTL, INC.
L & M ENGINEERING SOUTH
ELECTRONIC RESEARCH LABS
LOGIMETRICS, INC.
PRECISION STANDARD TIME, INC.
RF PRODUCTS, INC.
SHOKAI FAR EAST LTD.
SIGNATURE ANTENNA
SINCLAIR RADIO LABS INC.
SNYDER COMMUNICATIONS CORP.
TECH COMM. INC.
TEST EQUIPMENT CORP.
TRANS WORLD COMMUNICATIONS INC.
WATKINS-JOHNSON CO.

VHF

A.H. SYSTEMS, INC.
ADAMS-RUSSELL ELECTRONICS CO., INC.
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ALPHA INDUSTRIES, INC.
AMPLIFIER RESEARCH
ANDREW CORP.
ANTENNA SPECIALISTS CO., THE
CENTURION INTL INC.
CHILDS CORP.
COMAD COMMUNICATIONS LTD.
CUSHCRAFT CORP.
DAJEM TECHNOLOGY CO.
DAYTON-GRANGER, INC.
DECIBEL PRODUCTS, INC.
DIELECTRIC COMMUNICATIONS
DYMA ENGINEERING
FG ENGINEERING CO.
FLAM & RUSSELL, INC.
INTERAD LTD.
JAMES MILLEN ELECTRONICS

JOSEPH ELECTRONICS INC.
L & M ENGINEERING SOUTH
LARSEN ELECTRONICS, INC.
ELECTRONIC RESEARCH LABS
LOGIMETRICS, INC.
LUNAR INDUSTRIES, INC.
MARK ANTENNA PRODUCTS, INC.
MICRO COMMUNICATIONS, INC.
MICROWAVE SPECIALTY CORP.
MICROWAVE TECHNIQUES
RADIATION SYSTEMS, INC. MARK ANTENNAS DIVISION
SCALA ELECTRONIC CORP.
SENSOR SYSTEMS, INC.
SHOKAI FAR EAST LTD.
SIMPLICITY TOOL CO.
SINCLAIR RADIO LABS INC.
SNYDER COMMUNICATIONS CORP.
TECH COMM. INC.
TECOM
TELEWAVE INC.
TELSAT
TEST EQUIPMENT CORP.
TEXSCAN INSTRUMENTS
TIL-TEK LTD.
TRANSCO PRODUCTS, INC.
WATKINS-JOHNSON CO.

UHF

A.H. SYSTEMS, INC.
ADAMS-RUSSELL ELECTRONICS CO., INC.
AEL DEFENSE CORP.
ALPHA INDUSTRIES, INC.
AMPLIFIER RESEARCH
ANDREW CORP.
ANTENNA SPECIALISTS CO., THE
CENTURION INTL INC.
CHILDS CORP.
COMAD COMMUNICATIONS LTD.
CUSHCRAFT CORP.
DAJEM TECHNOLOGY CO.
DATRON SYSTEMS
DAYTON-GRANGER, INC.
DECIBEL PRODUCTS, INC.
DIELECTRIC COMMUNICATIONS
DYMA ENGINEERING
FG ENGINEERING CO.
FLAM & RUSSELL, INC.
INTERAD LTD.
JOSEPH ELECTRONICS INC.
L & M ENGINEERING SOUTH
LARSEN ELECTRONICS, INC.
ELECTRONIC RESEARCH LABS
LOGIMETRICS, INC.
LUNAR INDUSTRIES, INC.
MARK ANTENNA PRODUCTS, INC.
MICRO COMMUNICATIONS, INC.
MICROWAVE TECHNIQUES
ONEIDA ELECTRONIC MFG CO. INC.
RADIATION SYSTEMS, INC. MARK ANTENNAS DIVISION
SCALA ELECTRONIC CORP.
SCIENTIFIC ATLANTA
SENSOR SYSTEMS, INC.
SHOKAI FAR EAST LTD.
SIMPLICITY TOOL CO.
SINCLAIR RADIO LABS INC.
SNYDER COMMUNICATIONS CORP.
TECH COMM. INC.
TECOM
TELEWAVE INC.
TELSAT
TEST EQUIPMENT CORP.
TIL-TEK LTD.
TRANSCO PRODUCTS, INC.
WATKINS-JOHNSON CO.

EMC/EMI

A.H. SYSTEMS, INC.
ALPHA INDUSTRIES, INC.
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ANDREW CORP.
C.K. CONSULTANTS, INC.
DOW INDUSTRIES
EATON AIL DIV.
EATON CORP., ELECTRONICS INSTRUMENTATION DIV.
ELECTRO-MECHANICS CO., THE (EMCO)
EQUITABLE/RENTELCO CORP.
HEWLETT-PACKARD CO.
INSTRUMENTS FOR INDUSTRY, INC.
ELECTRONIC RESEARCH LABS
MAGNETIC RADIATION LABORATORIES, INC.
RADIOMETRICS CORP.

Transmission Components

RADIOMETRICS MIDWEST CORP.
SHOKAI FAR EAST LTD.
TEST EQUIPMENT CORP.
TEXAS SPECTRUM ELECTRONICS INC.
TIL-TEK LTD.
WATKINS-JOHNSON CO.

BROADBAND

ADAMS-RUSSELL ELECTRONICS CO., INC.
ALPHA INDUSTRIES, INC.
AMPLIFIER RESEARCH
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CENTURION INTL INC.
CHILDS CORP.
COMAD COMMUNICATIONS LTD.
CUBIC COMMUNICATIONS, INC.
CUSHCRAFT CORP.
DAJEM TECHNOLOGY CO.
DAYTON-GRANGER, INC.
DECIBEL PRODUCTS, INC.
DYMA ENGINEERING
HY GAIN
JOSEPH ELECTRONICS INC.
L & M ENGINEERING SOUTH
ELECTRONIC RESEARCH LABS
LOGIMETRICS, INC.
MARK ANTENNA PRODUCTS, INC.
MICRO COMMUNICATIONS, INC.
MICROFLECT CO., INC.
MICROWAVE SPECIALTY CORP.
NURAD, INC.
O.K. INDUSTRIES INC.
OLEKTRON CORP.
RADIATION SYSTEMS, INC. MARK ANTENNAS DIVISION
RADIOMETRICS MIDWEST CORP.
SCALA ELECTRONIC CORP.
SENSOR SYSTEMS, INC.
SHOKAI FAR EAST LTD.
SINCLAIR RADIO LABS INC.
SNYDER COMMUNICATIONS CORP.
SUNAIR ELECTRONICS
TECOM
TELEDYNE RYAN
TEST EQUIPMENT CORP.
TEXSCAN INSTRUMENTS
TIL-TEK LTD.
TRANSCO PRODUCTS, INC.
WATKINS-JOHNSON CO.

ANTENNA POSITIONERS

A.H. SYSTEMS, INC.
ANDREW CORP.
C.K. CONSULTANTS, INC.
DATRON SYSTEMS
DYMA ENGINEERING
DYNETIC SYSTEMS CO.
ELECTRO-MECHANICS CO., THE (EMCO)
FG ENGINEERING CO.
FLAM & RUSSELL, INC.
HUGHES AIRCRAFT CO. MICROWAVE PRODUCTS DIV.

**For More Information,
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HUGHES

Subsidiary of GM Hughes Electronics

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ROTATING PRECISION MECHANISMS, INC.
TIW SYSTEMS, INC.

VITRO SERVICES CORP.
WATKINS-JOHNSON CO.

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CALREP
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ESCO ASSOC.
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PICOSECOND PULSE LABS, INC.
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TEST PROBES, INC.
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TUCKER ELECTRONICS
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WAVETEK RF PRODUCTS, INC.
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WIDE BAND ENGINEERING CO. INC.
WILTRON CO.

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ELCOM SYSTEMS, INC.

ELECTRONIC RESOURCES INC.
ION BEAM MILLING, INC.
JFW INDUSTRIES, INC.
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RCD COMPONENTS INC.
REMEC
RLC ELECTRONICS, INC.
SALISBURY ENGINEERING INC.
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TELE-TECH CORP.
TEST EQUIPMENT CORP.
TEXSCAN INSTRUMENTS
WAVETEK RF PRODUCTS, INC.

PIN

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AMPLIFONIX, INC.
ANAREN MICROWAVE INC.
APPLIED SPECIALTIES, INC.
AYDIN VECTOR DIV.
CASTLE MICROWAVE
DAICO INDUSTRIES
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EATON CORP. MICROWAVE PRODUCTS DIV.
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ELECTROWAVE PRODUCTS, INC.
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JCA TECHNOLOGY INC.
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ELECTRONIC RESEARCH LABS
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MERRIMAC INDUSTRIES, INC.
MINI CIRCUITS
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OAKBURY COMPONENTS
OLEKTRON SYSTEMS, INC.
OPTIMAX, INC.
PASCALL ELECTRONICS LTD.
PENSTOCK INC.
PHOENIX MICROWAVE CORP.
PLANAR MICROWAVE, A RAYTHEON CO.
PREMIER MICROWAVE CORP.
QUAL-TECH, DIV. OF FLEXCO MICROWAVE INC.
REMEC
RLC ELECTRONICS, INC.
SDI MICROWAVE INC.
SHOKAI FAR EAST LTD.
SOLITRON DEVICES, INC.
TEKTRON MICRO ELECTRONICS INC.
TELE-TECH CORP.
TEST EQUIPMENT CORP.
TEXSCAN INSTRUMENTS
UNITRODE
VENTRONICS MICROWAVE CORP.
WATKINS-JOHNSON CO.

VARIABLE

ACCUTEST INSTRUMENTS INC.
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ADAMS-RUSSELL ELECTRONICS
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AIRTRON CORP.
ALAN INDUSTRIES, INC.
ALPHA INDUSTRIES, INC.
AMPLIFONIX, INC.
ANAREN MICROWAVE INC.
APOLLO MICROWAVES LTD.
APPLIED SPECIALTIES, INC.
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DAJEM TECHNOLOGY CO.
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ELECTRONIC RESOURCES INC.
ELECTROWAVE PRODUCTS, INC.
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HARRIS CORP./GSSD/PRD INSTRUMENTS
HEWLETT-PACKARD CO.

JFW INDUSTRIES, INC.
 KAY ELEMETRICS CORP.
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 ELECTRONIC RESEARCH LABS
 LORCH ELECTRONICS DIV. OF VERNITRON CORP.
 M/A-COM COMPONENTS MARKETING, INC.
 MARCH MICROWAVE LTD.
 MERRIMAC INDUSTRIES, INC.
 MICON, INC./MICROWAVE CONTROL CO. DIV.
 MICROLAB/FXR
 MICROWAVE TECHNIQUES
 MITEC ELECTRONICS LTD.
 NARDA MICROWAVE CORP.
 NORSAL INDUSTRIES INC.
 OLEKTRON SYSTEMS, INC.
 OPTIMAX, INC.
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 QUAL-TECH, DIV. OF FLEXCO MICROWAVE INC.
 RADIAL, INC.
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 REMEC
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 RLC ELECTRONICS, INC.
 ROBINSON LABS
 SHOKAI FAR EAST LTD.
 SOLITRON DEVICES, INC.
 SYSTRON DONNER CORP., MICROWAVE DIV.
 TACHONICS CORP.
 TELE-TECH CORP.
 TELONIC BERKELEY, INC.
 TEST EQUIPMENT CORP.
 TEXSCAN
 TEXSCAN INSTRUMENTS
 TONY CHAPMAN ELECTRONICS LTD.
 TUCKER ELECTRONICS
 WATKINS-JOHNSON CO.
 WAVELINE INC.
 WEINSCHTEL ENGINEERING

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 ALPHA INDUSTRIES, INC.
 ANGHEL LABORATORIES, INC.
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 ELECTRONIC RESOURCES INC.
 EMC TECHNOLOGY, INC.
 EQUITABLE/RENTELCO CORP.
 GENERAL MICROWAVE CORP.
 HARRIS CORP./GSSD/PRD INSTRUMENTS
 HEWLETT PACKARD CO., COMPONENTS
 HEWLETT-PACKARD CO.
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 OPTIMAX, INC.
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 PENSTOCK INC.
 QUAL-TECH, DIV. OF FLEXCO MICROWAVE INC.
 RADIAL, INC.
 REMEC
 RLC ELECTRONICS, INC.
 SHOKAI FAR EAST LTD.
 SOLITRON DEVICES, INC.
 SOLITRON/MICROWAVE
 TACHONICS CORP.
 TEKTRON MICRO ELECTRONICS INC.
 TEKTRONIX, INC.
 TELONIC BERKELEY, INC.
 TEXSCAN INSTRUMENTS
 TONY CHAPMAN ELECTRONICS LTD.
 TUCKER ELECTRONICS
 VENTRONICS MICROWAVE CORP.
 WAVETEK RF PRODUCTS, INC.
 WEINSCHTEL ENGINEERING
 WILTRON CO.

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 AIRTRON CORP.
 ALAN INDUSTRIES, INC.
 ALPHA INDUSTRIES, INC.
 AMPLIFONIX, INC.
 APPLIED SPECIALTIES, INC.
 DAICO INDUSTRIES
 DENSTRON MICROWAVE LTD.
 EATON CORP. MICROWAVE PRODUCTS DIV.
 ELECTRO RENT CORPORATION
 ELECTRONIC RESOURCES INC.
 EMC TECHNOLOGY, INC.
 EQUITABLE/RENTELCO CORP.
 GENERAL MICROWAVE CORP.
 HEWLETT-PACKARD CO.
 HUGHES AIRCRAFT CO. MICROWAVE PRODUCTS DIV.
 JFW INDUSTRIES, INC.
 KAY ELEMETRICS CORP.
 KDI ELECTRONICS
 ELECTRONIC RESEARCH LABS
 MARCH MICROWAVE LTD.
 MICROLAB/FXR
 MIDWEST MICROWAVE INC.
 NARDA MICROWAVE CORP.
 NORSAL INDUSTRIES INC.
 OPTIMAX, INC.
 PENSTOCK INC.
 PHOENIX MICROWAVE CORP.
 RADIAL, INC.
 RHG ELECTRONICS LABORATORY
 RLC ELECTRONICS, INC.
 SHOKAI FAR EAST LTD.
 TEKTRON MICRO ELECTRONICS INC.
 TELONIC BERKELEY, INC.
 TEST EQUIPMENT CORP.
 TEXSCAN INSTRUMENTS
 TONY CHAPMAN ELECTRONICS LTD.
 TUCKER ELECTRONICS
 WAVELINE INC.
 WAVETEK RF PRODUCTS, INC.
 WEINSCHTEL ENGINEERING
 WILTRON CO.

HIGH POWER

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 AEROWAVE INC.
 AIRTRON CORP.
 ALAN INDUSTRIES, INC.
 ALPHA INDUSTRIES, INC.
 APOLLO MICROWAVES LTD.
 BIRD ELECTRONIC CORP.
 EMC TECHNOLOGY, INC.
 EQUITABLE/RENTELCO CORP.
 INMET CORP.
 JFW INDUSTRIES, INC.
 KAY ELEMETRICS CORP.
 KDI ELECTRONICS
 ELECTRONIC RESEARCH LABS
 MICON, INC./MICROWAVE CONTROL CO. DIV.
 MICROLAB/FXR
 MICROWAVE TECHNIQUES
 NARDA MICROWAVE CORP.
 PASTERNAK ENTERPRISES

PENSTOCK INC.
 POLYMER DESIGN
 QUAL-TECH, DIV. OF FLEXCO MICROWAVE INC.
 RAYTHEON CO., MPTD
 RLC ELECTRONICS, INC.
 SHOKAI FAR EAST LTD.
 SOLITRON DEVICES, INC.
 TEST EQUIPMENT CORP.
 TEXSCAN INSTRUMENTS
 WAVETEK RF PRODUCTS, INC.
 WEINSCHTEL ENGINEERING

COAXIAL CABLE

ADAMS-RUSSELL ELECTRONICS
 ADAMS-RUSSELL ELECTRONICS CO., INC.
 AMPHENOL NORTH AMERICA
 AMPHENOL RF/MICROWAVE OPERATIONS
 ANDREW CORP.
 ANTENNA SPECIALISTS CO., THE
 APPLIED SPECIALTIES, INC.
 BELDEN WIRE AND CABLE
 BI-TRONICS, INC.
 BRAND REX CABLE SYSTEMS
 BREEZE ILLINOIS
 BRIM ELECTRONICS, INC.
 C & M CORP.
 CALREP
 CAPCON, INC.
 CASTLE MICROWAVE
 CERAMSEAL
 COAXIAL COMPONENTS CORP. (COAXICOM)
 COAXITUBE CORP.
 COMARK COMMUNICATIONS, INC.
 CONNECTING DEVICES, INC.
 CUSHCRAFT CORP.
 DABURN ELECTRONICS & CABLE CORP.
 DECIBEL PRODUCTS, INC.
 DIELECTRIC COMMUNICATIONS
 DIELECTRIC SCIENCES
 DYMA ENGINEERING
 ELECTRONIC RESOURCES INC.
 ESCO ASSOC.
 FLEXCO MICROWAVE, INC.
 FM&T INDUSTRIES, INC.
 FREDERICK COMPONENTS
 GILBERT ENGINEERING
 GILTRON INC.
 GLENAIR CORP.
 HABIA CABLE
 HARBOUR INDUSTRIES, INC.
 HAVERHILL CABLE & MFG CORP.
 HUBER & SUHNER INC.
 INSULATED WIRE INC.
 ITT POMONA ELECTRONICS
 ITT SCHADOW INC.
 JOSEPH ELECTRONICS INC.
 JOSLYN DEFENSE SYSTEMS, INC.
 KAMAN INSTRUMENTATION CORP.
 KINGS ELECTRONICS CO. INC.
 L & M ENGINEERING SOUTH
 LANCE WIRE AND CABLE, INC.
 LARSEN ELECTRONICS, INC.
 ELECTRONIC RESEARCH LABS
 MADISON CABLE CORP.
 MALCO, MICRODOT INC., CONNECTOR, INTERCON-
 NECT SYSTEMS DIV.
 MANHATTAN ELEC. CABLE CORP.
 MARCH MICROWAVE LTD.
 MARSHALL ELECTRONICS
 MICRO COAX COMPONENTS, INC.
 MIDISCO
 MINELCO INC.
 MOLEX INC.
 NEK CABLE, INC.
 PASCALL ELECTRONICS LTD.
 PASTERNAK ENTERPRISES
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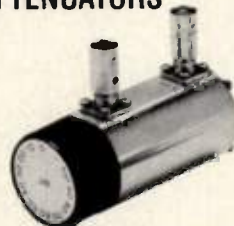
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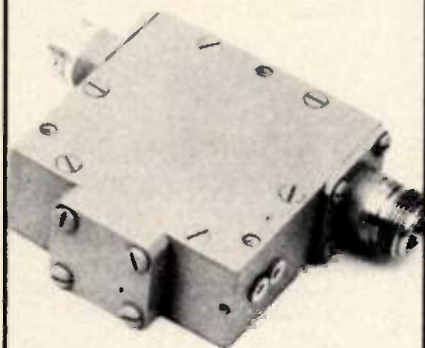
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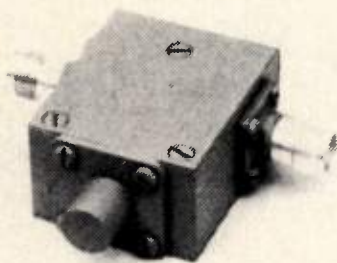
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- FAST SWITCHING DIODES
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- MIS CHIP CAPACITORS

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- DROP-INS
- MICROSTRIP CHIP
- MULTI-OCTAVE
- OPTIMIZED STANDARD BANDS TO 21 GHz

GAIN EQUALIZERS

- ADJUSTABLE OR FIXED CONFIGURATIONS
- BALL BEARING DRIVE AND CONTACT
- FREQUENCY AND AMPLITUDE TUNABLE
- SYSTEM OR TWT APPLICATIONS
- HARMONIC PHASE ENHANCERS

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- SCHOTTKY TUNNEL DETECTORS
- LIMITER – DETECTORS
- HIGH POWER LIMITERS
- ISOTECTORS & COUPLER DETECTORS
- ZERO BIAS DETECTORS/POWER MONITORS

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	U 862 BS	2.4	40	250 mv	T050
	U 822 BS	2.1	31	250 mv	T050
	U 2620 BS	2.1	30	250 mv	T050 Ceramic
4	U 864 BS	2.4	45	250 mv	T050
	U 824 BS	2.1	37	250 mv	T050
	U 4620 BS	2.1	35	250 mv	T050 Ceramic
8	U 6608 B	6.0			T050
64	U 664 BS	1.0	50	ECL	DIP8
	U 684 BS	1.0	32	ECL	DIP8, SIP6, SO8
64,128, 256	U 833 BS	1.3	45	ECL	DIP8, SIP6, SO8
	U 833 BSE	1.3	45	EF	DIP8, SIP6, SO8
	U 813 BS	1.1	35	ECL	DIP8, SIP6, SO8
	U 813 BSE	1.1	35	EF	DIP8, SIP6, SO8
256	U 656 BS	1.5	30	ECL	DIP8
	U 847 BST	1.3	TBE	TTL	DIP8
	U 666 B	1.0	53	ECL	DIP8, SIP6
	U 666 BS	1.0	65	TTL	DIP8, SIP6
	U 686 BS	1.0	32	ECL	DIP8, SIP6, SO8
4096	U 496 BS	1.0	60	TTL	DIP8

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Tuning**

1.0 GHz	PLL	U 6316 B
500 MHz	Front End	U 4062 B
470 MHz	Front End	U 4777 B
200 MHz	Saw Driver	U 4744 B
1.3 GHz	PLL	U 6202 B

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Amplifiers**

1.8 GHz	U 4518 B
2.0GHz	U 4521 B

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Creative Technologies

INFO/CARD 50

HIGH POWER COMBINERS



**2-WAY 600 WATTS
MODEL D1994**

**4-WAY 600 WATTS
MODEL D1995**

**4-WAY 1500 WATTS
MODEL D1996**

SPECIFICATIONS above models

Freq. Range	20-90 MHz
Insertion Loss	0.3 dB max.
Isolation	20 dB min.
VSWR	1.3:1 max.

HIGH POWER COUPLERS

**DUAL DIRECTIONAL
COUPLER
MODEL C1997**

Freq. Range	20-90 MHz
Coupling	40 dB
Directivity	20 dB min.
Power (watts)	1500

See Gold Book and Microwave/RF Handbook, for additional products.



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BREWSTER, NY 10509
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TWX 510-600-0747

INFO/CARD 51

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See Matrix Systems Ad on Coaxial Switching

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INFO/CARD 52

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See Matrix Systems Ad on Coaxial Switching

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INFO/CARD 53

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INFO/CARD 55

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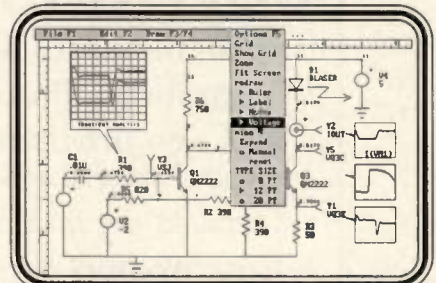
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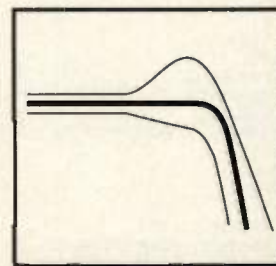
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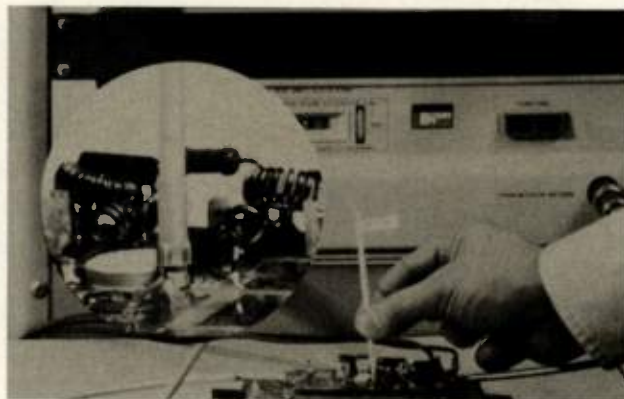
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68	
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470	
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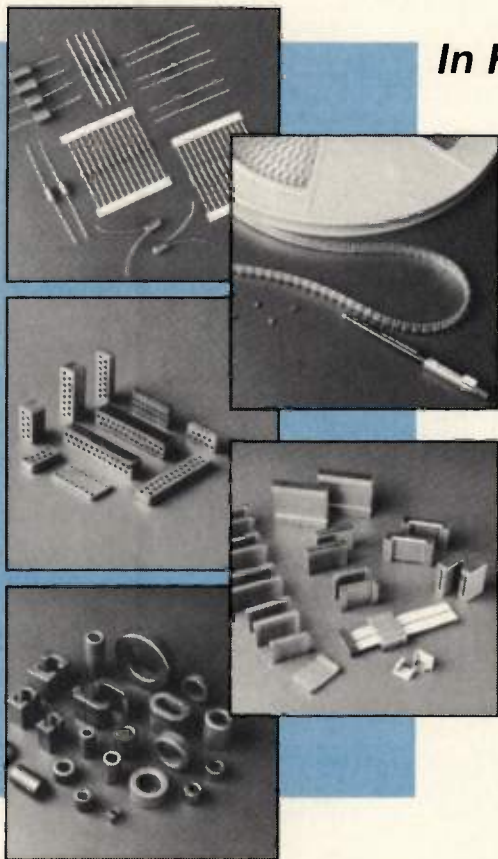
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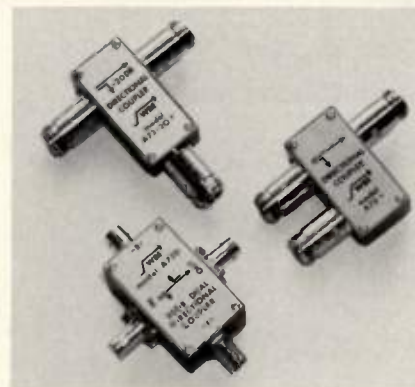
APPLICATIONS

Some general applications for the A73 Series are:

Line Monitoring: Power split from the line is -20 dB down for sampling without altering line characteristics, for level measuring, VSWR alarms, etc..

Power Measurements: Insertion in the line allows level measurements with simple lower level detectors or field strength meters and power measuring equipment. By reversing the coupler in the line or using the A73D types, an indication of impedance match and/or reflected power can be measured by comparing the forward to reflected power levels.

Load Source Isolator: Using a directional coupler in the line, a signal can be taken from the source to the tap with high attenuation (directivity) between the tap and the load.



Model	Freq. Range MHz	Coupling Level dB	Coupler Type	In Line Power	Minimum Directivity 1-500 (dB) 5-300 MHz	In Line Loss (dB)	Flatness of Coupled Port (dB)	VSWR	Price 50 ohm with BNC conns.
A73-20	1-500	20	single	5W cw (10W cw 5-300 MHz)	20	.4 max. .2 typical	±.1 5-300 MHz ±.25 1-500 MHz	1.05:1	\$ 68.00
A73-20GA					30			5-500 MHz	131.00
A73-20GB					40			1-500 MHz	242.00
A73-20P	1-100	20	single	50W cw	35 dB min. 40 dB min. typical	.15	±.1	1.1:1 max	91.00
A73D-20P			dual			.3			163.00
A73-20PAX	10-200	20	single	(75 ohm limited to 10 W cw)	45 dB min.	.15	±.1	1.04:1 typical	150.00
A73D-20PAX			dual			.3			310.00
A73-30P2	1-100	30	single	200 W cw 50 ohm	30 dB	.05	±.15	1.05:1 max	312.00

This chart is just a sampling of couplers available. Connector options are available. Consult factory for specials and OEM applications.

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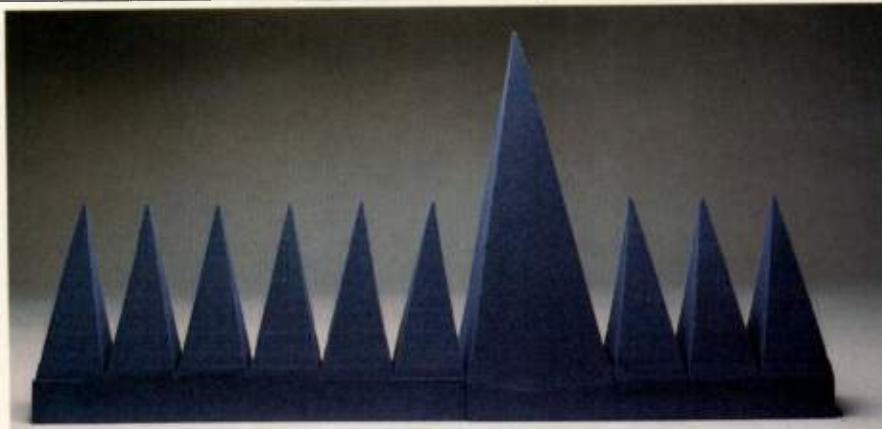
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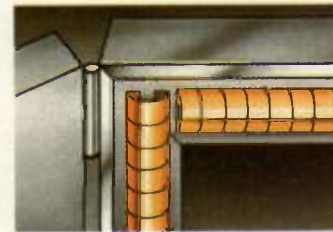
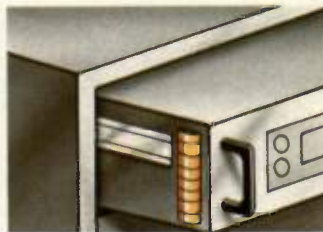
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Low Impedance Double-Tuned Matching Circuit

By Andrzej B. Przedpelski
A.R.F. Products, Inc.

Filters at 50-300 ohm impedance levels are usually hard to design because of component value limitations. A circuit which overcomes this problem is the double-tuned circuit with taps on the input and output. Real input and output impedances are assumed. Any stray capacitance can be subtracted from C_1 and C_2 values.

The required inputs are (steps 80-130):

F_0 = center frequency, MHz
 BW = approximate bandwidth, MHz
 Q_L = inductor Q
 R_s = source resistance, ohms
 R_o = load resistance, ohms
 L = inductance, uH

Using the values shown in the program, the circuit values are:

$X_L = 188.47$	
$X_1 = -31.604$	$C_1 = 167.89 \text{ pF}$
$X_2 = -161.66$	$C_2 = 32.821 \text{ pF}$
$X_3 = -8076.2$	$C_3 = 0.65698 \text{ pF}$
$X_4 = -180.52$	$C_4 = 29.392 \text{ pF}$
$X_5 = -13.276$	$C_5 = 399.66 \text{ pF}$

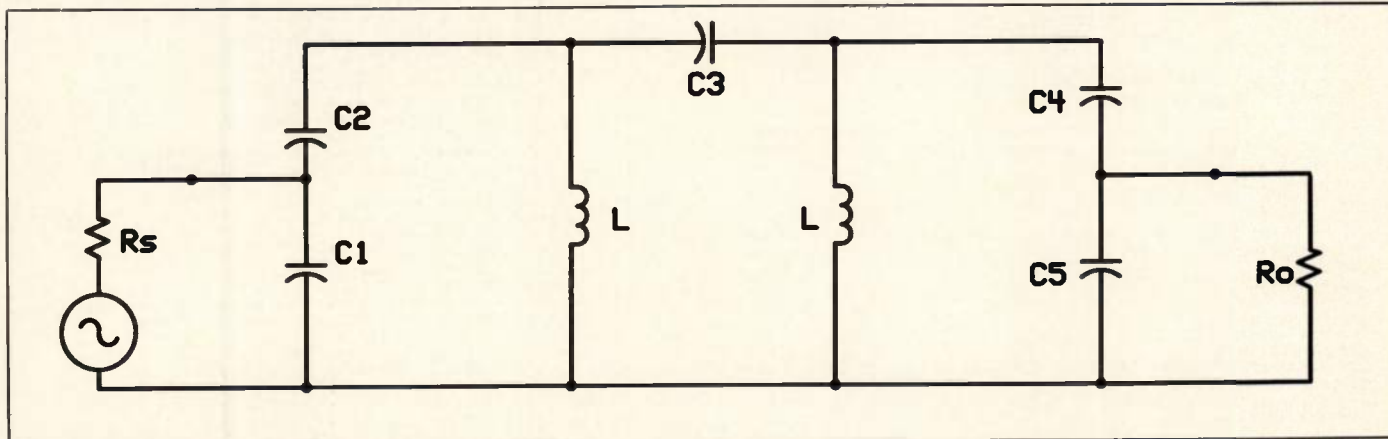
About the Author

Andrzej B. Przedpelski is vice president, development of A.R.F. Products, Inc. 2559 75th St., Boulder, CO 80301. He serves as consulting editor to *RF Design*.

```

10 ! LOW IMPDANCE DOUBLE TUNED MATCHING CIRCUIT - #595
20 ! A. PRZEDPELSKI - 18 APRIL 1988
30 ! A.R.F. PRODUCTS, INC. LI DTUNE"
40 ! BOULDER, CO
50 ! REF: "Low Impedance Double Tuned Circuits"
60 ! RF DESIGN - MAY/JUN 1982
70 ! ***** ENTER KNOWN *****
80 F0=30 ! Center frequency, MHz
90 BW=1 ! Approximate desired bandwidth, MHz
100 QL=150 ! Inductor Q
110 RS=300 ! Source (series) resistance, ohm
120 RO=50 ! Load (series) resistance, ohm
130 L=1 ! Inductor, uH
140 ! ***** CALCULATION *****
150 SHORT C1,C2,C3,C4,C5,XL,X1,X2,X3,X4,X5
160 F=SQR ((F0-.5*BW)*(F0+.5*BW))*1000000 ! Geometric center frequency
170 W=2*PI *F
180 Q0=F/(.7*BW*1000000)
190 XL=W*L*.000001
200 RL=QL*XL
210 X3=-(XL*Q0)
220 C3=-(1/(W*X3))*1.E12 ! IN pF
230 A1=(-X3)*RL/(RL+X3)
240 A2=X3/(Q0-1)
250 A3=(A1/A2)^2
260 A4=A1/(1+A3)
270 X1=-SQR (RS^2+A4/(RS-A4))
280 C1=-(1/(W*X1))*1.E12 ! IN pF
290 X5=-SQR (RO^2+A4/(RO-A4))
300 C5=-(1/(W*X5))*1.E12 ! IN pF
310 A5=A2*A3/(1+A3)
320 A6=(A5/X1)^2
330 X2=A5-X1*((RS/X1)^2/(1+(RS/X1)^2))
340 C2=-(1/(W*X2))*1.E12 ! IN pF
350 X4=A5-X5*((RO/X5)^2/(1+(RO/X5)^2))
360 C4=-(1/(W*X4))*1.E12 ! IN pF
370 CLEAR @ DISP "CENTER FREQUENCY = ";F0;"MHz"
380 DISP "BANDWIDTH = ";BW;"MHz"
390 DISP "COIL Q = ";QL
400 DISP "XL = ";XL,"L = ";L;"uH"
410 DISP "X1 = ";X1,"C1 = ";C1;"pF"
420 DISP "X2 = ";X2,"C2 = ";C2;"pF"
430 DISP "X3 = ";X3,"C3 = ";C3;"pF"
440 DISP "X4 = ";X4,"C4 = ";C4;"pF"
450 DISP "X5 = ";X5,"C5 = ";C5;"pF"
460 END

```



Double-tuned circuit with taps on input and output.

Tapped Double-Tuned Matching Circuit (Bandpass)

By Andrzej B. Przedpelski
A.R.F. Products, Inc.

When more selectivity is desired than a simple matching circuit can provide, and a considerable impedance step-up/step-down is needed, the double-tuned tapped circuit can be useful. A simple double-tuned circuit relies on the mutual coupling between two tuned circuits. For optimum performance, critical coupling is normally used. In conventional IF transformers, the coupling is usually inductive, but for most applications, especially R&D work, the capacitively coupled circuit is easier to design and adjust. Providing a capacitive tap on one side permits a large input/output impedance ratio.

The needed inputs are:

R_s = source resistance, ohms
 P_s = source capacitance or inductance
 (- for capacitance), pF or uH
 R_o = load resistance
 (must be less than R_s), ohms
 P_o = load capacitance or inductance
 (- for capacitance), pF or uH
 Q_0 = circuit Q
 (can be selected to obtain the desired bandwidth, must be less than Q_L)

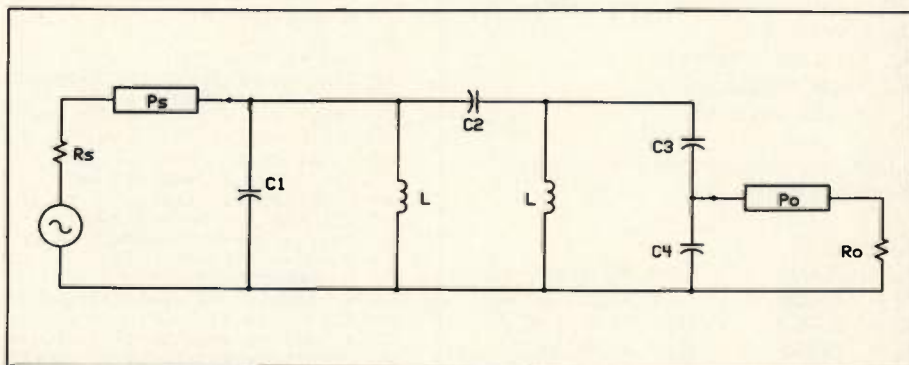
Q_L = inductor Q
 (it is assumed that capacitor Q_s
 are high)
 F_0 = center frequency, Hz

These values are overwritten over the values shown in the program in steps 80-140. If desired, these can be changed to an *Input* statement. The program provides all the necessary circuit values. It also provides an approximate value of

loss caused by the finite inductor Q.

Using the values shown in the program, the following circuit elements were obtained:

$L = 0.53649 \text{ uH}$
 $C_1 = 124.04 \text{ pF}$
 $C_2 = 6.5576 \text{ pF}$
 $C_3 = 181.4 \text{ pF}$
 $C_4 = 417.67 \text{ pF}$



The double-tuned tapped circuit.

```

10 ! TAPPED DOUBLE-TUNED MATCHING CIRCUIT (BANDPASS)
11 ! PROGRAM NO. 592 - "DTUNE"
12 ! A. PRZEDPELSKI - 15 APRIL 1988
13 !
14 ! REF.: "Double Tuned Circuits"
15 ! RF DESIGN - JAN/FEB 1982
16 ! ***** ENTER KNOWN *****
17 F0=60000000 ! Frequency, Hz
18 Q0=20 ! Desired circuit Q
19 QL=100 ! Inductor Q
20 Rs=500 ! Source (series) resistance, ohm
21 Ps=-50 ! Source capacity or inductance (- for capacity), pF or uH
22 Ro=50 ! Load (series) resistance, ohm
23 Po=.05 ! Load capacity or inductance (- for capacity), pF or uH
24 ! ***** CALCULATION *****
25 SHORT L,C1,C2,LOSS,C3,C4
26 W=2*PI *F0
27 IF Ps<0 THEN Xs=1/(W*Ps*.000000000001) ELSE Xs=W*Ps*.000001
28 IF Po<0 THEN Xo=1/(W*Po*.000000000001) ELSE Xo=W*Po*.000001
29 RS=Rs @ XS=Xs @ GOSUB STOR
30 R1=RP @ X5=XP
31 RS=Ro @ XS=Xo @ GOSUB STOR
32 R5=RS @ X10=XP
33 IF R1<R5 THEN DISP "REVERSE COUPLING CIRCUIT" @ LIST 70,140
34 XL=R1*(Q0-Q0)/(Q0-Q0)
35 L=XL/W
36 RL=QL/XL
37 Rp=Q0/XL
38 R2=RL*Rp/(RL-Rp)
39 X2=-(Q0*XL)
40 C2=-(1/(W*X2))
41 X1=X2*X5/(X5*(Q0-1)-X2)
42 C1=-(1/(W*X1))
43 X6=1/(1/X1+1/X5)
44 X7=1/(1/X6+1/XL)
45 K=1/Q0
46 X6=X2/(1/K-1)
47 X7=1/(1/X6+1/XL)
48 RP=R1 @ XP=X6 @ GOSUB PTOS
49 R6=RS @ X13=X5
50 X3=X13+R6*SQR (R5/R6-1)
51 C3=-(1/(W*X3))
52 R6=RS @ X11=X13-X3
53 X4=X11*X10/(X10-X11)
54 C4=-(1/(W*X4))
55 DISP "L = ":L*1000000:"uH"
56 DISP "C1 = ":C1*1.E12:"pF"
57 DISP "C2 = ":C2*1.E12:"pF"
58 DISP "C3 = ":C3*1.E12:"pF"
59 DISP "C4 = ":C4*1.E12:"pF"
60 LOSS=2*(-(10*LGT (1-Q0/QL)))
61 DISP "AFRDX. LOSS (due to QL) = ":LOSS:"dB"
62 END
63 ! ***** STANDARD SUBROUTINES *****
64 STOR: ! SERIES TO PARALLEL CONVERSION
65 ! INPUT RS, XS
66 IF RS=0 THEN RS=1.E-30
67 IF XS=0 THEN XS=1.E-30
68 RP=RS^2/RS+XS
69 XP=RS^2/XS+XS
70 ! OUTPUT RP, XP
71 RETURN
72 PTOS: ! PARALLEL TO SERIES CONVERSION
73 ! INPUT RP, XP
74 IF XP=0 THEN XP=1.E-30
75 RS=RP/(1+(RP/XP)^2)
76 XS=XP*RP^2/(XP^2+RP^2)
77 ! OUTPUT RS, XS
78 RETURN

```

Response of Low Impedance Double-Tuned Matching Circuit

By Andrzej B. Przedpelski
A.R.F. Products, Inc.

The frequency response of the low impedance double-tuned matching circuit can be calculated using this program. The circuit values are entered in steps 90 to 170. The program will then ask for start frequency, end frequency and frequency steps in MHz. It then prints out a frequency response table in dB. In step 1170, if the ATN2 function is unavailable (which places it in the quadrant), the equivalent subroutine may have to be substituted. The remarks refer to the equations and variables in the referenced article. Using the given circuit values, the table shown was obtained using 0.5 MHz steps.

Note: The response (insertion loss) includes loss due to the finite inductor Q and the voltage step-down due to impedance step-down.

References

1. Andrzej B. Przedpelski, "Low Impedance Double Tuned Circuit," *RF Design*, May/June 1982.

FREQUENCY, MHz	RESPONSE, dB	PHASE, DEG	GAMMA	FREQUENCY, MHz	RESPONSE, dB	PHASE, DEG	
25	-60.215	-89.036	.99799	30.5	-19.604	-58.482	.57714
25.5	-58.011	-88.627	.99751	31	-28.568	-82.794	.92019
26	-55.601	-87.082	.99682	31.5	-35.11	-86.641	.97106
26.5	-52.927	-86.251	.99581	32	-39.792	-87.953	.9852
27	-49.904	-80.853	.99421	32.5	-43.387	-88.583	.99099
27.5	-46.402	12.521	.99146	33	-46.285	-88.946	.99393
28	-42.204	76.359	.9861	33.5	-48.7	-89.177	.99562
28.5	-36.917	79.719	.97313	34	-50.763	-89.336	.99669
29	-29.748	77.214	.92657	34.5	-52.557	-89.45	.99741
29.5	-19.976	54.548	.59552	35	-54.141	-89.535	.99792
30	-16.724	-6.2611	.28574				

Frequency response calculated from 25 to 35 MHz.

```

10 ! RESPONSE OF LOW IMPEDANCE DOUBLE TUNED MATCHING CIRCUIT - #596
20 ! A. PRZEDPELSKI - 19 APRIL 1988
30 ! A.R.F. PRODUCTS, INC.
40 ! BOULDER, CO
50 ! REF.: "Low Impedance Double Tuned Circuits"
60 ! RF DESIGN - MAY/JUN 1982
70 ! Program No. 595 calculates circuit values
80 ! ***** ENTER KNOWNS *****
90 R=300 ! Source (series) resistance, ohm
100 Ro=50 ! Load (series) resistance, ohm
110 L=1 ! Inductor, uH
120 QL=150 ! Inductor Q
130 C1=167.89 ! Parallel input capacitor, pF
140 C2=32.821 ! Series input capacitor, pF
150 C3=.65698 ! Coupling capacitor, pF
160 C4=29.392 ! Series output capacitor, pF
170 C5=399.66 ! Series output capacitor, pF
180 ! ***** CALCULATION *****
190 CLEAR @ DEG @ SHORT LM,LP,GAIN,GAINM,GAMMA
200 DISP "START FREQUENCY, MHz = " @ INPUT F1
210 DISP "STOP FREQUENCY, MHz = " @ INPUT F2
220 DISP "FREQUENCY STEPS, MHz = " @ INPUT FS
230 CLEAR
240 DISP "FREQUENCY, MHz", "RESPONSE, dB", "PHASE, DEG", "GAMMA" @ DISP
250 FOR F=F1 TO F2 STEP FS
260 W=2*PI*F*1000000
270 XL=W*L*.000001 ! Reactance of L
280 X1=-(1/(W*C1*.000000000001)) ! Reactance of C1
290 X2=-(1/(W*C2*.000000000001)) ! Reactance of C2
300 X3=-(1/(W*C3*.000000000001)) ! Reactance of C3
310 X4=-(1/(W*C4*.000000000001)) ! Reactance of C4
320 X5=-(1/(W*C5*.000000000001)) ! Reactance of C5
330 RL=QL*XL ! Close approximation
340 RP=Ro @ XP=X5
350 GOSUB PTOS
360 R1=RS @ X6=X5 ! (b)
370 X7=X6+X4 ! (c)
380 RS=R1 @ XS=X7
390 GOSUB STOR
400 R2=RP @ X8=XP ! (d)
410 X9=1/(1/X8+1/XL) @ R3=1/(1/R2+1/RL) ! (e)
420 RP=R3 @ XP=X9
430 GOSUB PTOS
440 R4=RS @ X10=X5 ! (f)
450 X11=X10+X3 ! (g)
460 RS=R4 @ XS=X11
470 GOSUB STOR
480 R5=RP @ X12=XP ! (h)
490 X13=1/(1/X12+1/XL) @ R6=1/(1/R5+1/RL) ! (i)
500 RP=R6 @ XP=X13
510 GOSUB PTOS
520 R7=RS @ X14=X5 ! (j)
530 X15=X14+X2 ! (k)
540 RS=R7 @ XS=X15
550 GOSUB STOR
560 R8=RP @ X16=XP ! (l)
570 X17=1/(1/X16+1/X1) ! (m)
580 RP=R8 @ XP=X17
590 GOSUB PTOS
600 R9=RS @ X18=X5 ! (n)
610 ! ***** GAMMA CALCULATION *****
620 R=R9-Rs @ X=X18 @ GOSUB RTOP
630 G1=MAG
640 R=R9+Rs @ X=X18 @ GOSUB RTOP
650 G2=MAG
660 GAMMA=G1/G2 ! Reflection coefficient
670 ! ***** RESPONSE CALCULATION *****
680 R=1 @ X=X0 @ GOSUB RTOP
690 E1=MAG @ E1A=ANG ! Input voltage in polar form (e1=1)
700 R=Rs+R9 @ X=X18 @ GOSUB RTOP
710 Z1=MAG @ Z1A=ANG ! Total input impedance (including source)
720 I1=E1M/Z1M @ I1A=E1A-Z1A ! Input current in polar form
730 R=R9 @ X=X18 @ GOSUB RTOP
740 NM=MAG @ NA=ANG ! Numerator of (26) in polar form
750 R=R7 @ X=X15 @ GOSUB RTOP
760 DM=MAG @ DA=ANG ! Denominator of (26) in polar form
770 I3=I1M*NM/DM @ I3A=I1A+NA-DA ! I3 in polar form
780 R=R7 @ X=X14 @ GOSUB RTOP
790 NM=MAG @ NA=ANG
800 R=R4 @ X=X11 @ GOSUB RTOP
810 DM=MAG @ DA=ANG
820 I5M=I3M*NM/DM @ I5A=I3A+NA-DA ! I5 in polar form
830 R=R4 @ X=X10 @ GOSUB RTOP
840 NM=MAG @ NA=ANG
850 R=R1 @ X=X7 @ GOSUB RTOP
860 DM=MAG @ DA=ANG
870 I7M=I5M*NM/DM @ I7A=I5A+NA-DA ! I7 in polar form
880 R=R1 @ X=X6 @ GOSUB RTOP
890 NM=MAG @ NA=ANG
900 DM=Ro @ DA=0
910 I9M=I7M*NM/DM @ I9A=I7A+NA-DA ! I9 in polar form
920 EoM=I9M*Ro @ EoA=I9A ! Output voltage in polar form
930 E1M=1 @ E1A=0 ! Input voltage in polar form
940 GAINM=20*LGT (EoM/E1M) @ GAINA=EoA-E1A ! Gain in polar form
950 ! ***** DATA DISPLAY *****
960 DISP F,GAINM,GAINA,GAMMA
970 NEXT F
980 END
990 ! ***** STANDARD SUBROUTINES *****
1000 STOR: ! Series to parallel conversion
1010 ! INPUT RS, XS
1020 IF RS=0 THEN RS=1.E-30
1030 IF XS=0 THEN XS=1.E-30
1040 RP=XS^2/RS+XS
1050 XP=RS^2/XS+XS
1060 ! OUTPUT RP, XP
1070 RETURN
1080 PTOS: ! Parallel to series conversion
1090 ! INPUT RP, XP
1100 IF XP=0 THEN XP=1.E-30
1110 RS=RP/(1+(RP/XP)^2)
1120 XS=XP*RP^2/(XP^2+RP^2)
1130 ! OUTPUT RS, XS
1140 RETURN
1150 RTOP: ! Rectangular to polar conversion
1160 ! INPUT R, X - series resistance and reactance
1170 MAG=SOR (R^2+X^2) @ ANG=ATN2 (XS,RS)
1180 ! OUTPUT MAG, ANG - magnitude and angle
1190 RETURN

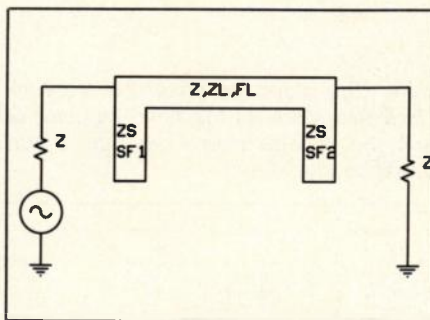
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Dual Trap Frequency Response

By Andrzej B. Pzredpelski
A.R.F. Products, Inc.

The dual trap circuit is useful when a single frequency or a small frequency band has to be eliminated. The circuit shown can be designed to provide a large variety of frequency responses. For instance, the two traps can be tuned to the same frequency to provide maximum single frequency rejection or the traps can be stagger tuned to provide a rejection band. The line between the open circuited traps adjusts the overall response characteristic. For instance, if both traps are tuned to the same frequency, and the connecting line is 90 or 180 degrees long (at the trap frequency), the response will be symmetrical.

On the other hand a very short line will give an overall skewed response, while the maximum frequency of attenuation will remain the same. The two traps are 90 degrees long at the desired maximum attenuation frequencies and are open circuited stubs. The stubs do not have to be



Dual trap circuit.

the same characteristic impedance as the system impedance, but can be adjusted for the desired frequency response or space and line impedance available.

An example is given for a narrow band trap with the first trap tuned to 99 MHz and the second to 101 MHz, the connecting line being 90 degrees at 100 MHz. All lines are 50 ohms.

FREQUENCY, HZ	GAIN, DB
90000000	-25.538
91000000	-27.52
92000000	-29.713
93000000	-32.178
94000000	-35.01
95000000	-38.362
96000000	-42.506
97000000	-48.014
98000000	-56.567
99000000	-1.E00
100000000	-66.137
101000000	-1.E00
102000000	-56.567
103000000	-48.014
104000000	-42.506
105000000	-38.362
106000000	-35.01
107000000	-32.177
108000000	-29.712
109000000	-27.52
110000000	-25.538

Frequency versus gain.

```

10 REM DUAL TRAP [OPEN STUB] RESPONSE - 0545
20 : A. PRZEDPELSKI - 24 FEB 1986 / 22 AUG 1988
30 : A.R.F. PRODUCTS, INC.
40 : BOULDER, CO
50 : REF.: "CALCULATOR DESIGN OF MICROSTRIP TRAPS"
60 : R.F. DESIGN, JULY/AUGUST 1979
70 CLEAR @ DEG @ SHORT L.F
80 DISP "SECOND TRAP FREQUENCY [HZ]:" @ INPUT SF2
90 DISP "FIRST TRAP FREQUENCY [HZ]:" @ INPUT SF1
100 DISP "STUB IMPEDANCES:" @ INPUT ZS
110 DISP "ANGLE AND FREQUENCY [HZ] OF LINE BETWEEN STUBS:" @ INPUT ZL,FL
120 DISP "START, STOP AND STEP FREQUENCIES [HZ]:" @ INPUT S1,S2,S3
130 DISP "SYSTEM CHARACTERISTIC IMPEDANCE:" @ INPUT Z
140 DISP @ DISP "FREQUENCY, HZ", "GAIN, DB"
150 : *****
160 FOR F=S1 TO S2 STEP S3 : FREQUENCY SWEEP
170 : ***** CALCULATE SECOND STUB *****
180 W2=F/SF2 : SECOND STUB ANGLE CORRECTION
190 Zo=ZS : SECOND STUB LINE IMPEDANCE
200 SANG=90 : LENGTH OF OPEN STUB
210 ANG=SANG+W2 : CORRECTED SECOND STUB LENGTH FOR FREQUENCY
220 Ro=1.E30 : OPEN CIRCUITED STUB
230 Xo=1.E30 : " "
240 : *****
250 GOSUB ZTRANS : CALCULATE SECOND STUB REACTANCE
260 XSTUB2=X1 : SECOND STUB REACTANCE
270 : ***** CALCULATE FIRST STUB *****
280 W1=F/SF1 : ANGLE CORRECTION
290 Zo=ZS : STUB LINE IMPEDANCE
300 ANG=SANG+W1 : CORRECTED SECOND STUB LENGTH FOR FREQUENCY
310 GOSUB ZTRANS : CALCULATE FIRST STUB REACTANCE
320 XSTUB1=X1 : SECOND STUB REACTANCE
330 : ***** CALCULATE IMPEDANCE AT SECOND STUB *****
340 RP=Z : PARALLEL OUTPUT RESISTANCE
350 XP=XSTUB2 : PARALLEL STUB REACTANCE
360 GOSUB PTOS : CONVERT TO SERIES IMPEDANCE
370 R2=RS @ X2=X5 : IMPEDANCE AT SECOND STUB
380 : ***** CALCULATE IMPEDANCE AT FIRST STUB *****
390 Ro=R2 @ Xo=X2 : IMPEDANCE AT SECOND STUB
400 W3=F/FL : CONNECTING LINE CORRECTION
410 ANG=W3*ZL : CONNECTING LINE LENGTH CORRECTED
420 Zo=Z : CONNECTING LINE IMPEDANCE
430 GOSUB ZTRANS : TRANSFORM TO FIRST STUB
440 RS=R1 @ XS=X1 : CONVERTED SECOND STUB IMPEDANCE AT FIRST STUB
450 GOSUB STOR : CONVERT TO PARALLEL
460 XF=1/(1/XSTUB1+1/XP) : TOTAL PARALLEL REACTANCE
470 GOSUB PTOS : CONVERT TO SERIES
480 R=RS @ X=X5 : SERIES IMPEDANCE AT FIRST STUB
490 : ***** CALCULATE INSERTION LOSS *****
500 GOSUB GAM : CALCULATE GAMMA AT FIRST STUB

510 GOSUB LOSS : CALCULATE LOSS
520 : ***** OUTPUT DATA *****
530 DISP F,L
540 NEXT F
550 END

560 REM ***** STANDARD ROUTINES *****
570 PTOS: : PARALLEL TO SERIES CONVERSION
580 : INPUT RP,XP
590 RS=RP/(1+(RP/XP)^2)
600 XS=XP*RP^2/(XF^2+RP^2)
610 : OUTPUT RS,XS
620 RETURN
630 ZTRANS: : TRANSFORMATION USING TRANSMISSION LINE
640 : INPUT Zo,ANG,Ro,Xo
650 IF ANG=90 THEN T=1.E100 ELSE T=TAN (ANG)
660 A=Zo*Ro @ B=Zo*Xo+Zo^2*T
670 C=Zo-Xo*T @ D=Ro*T
680 R1=(A+C*B*D)/(C^2+D^2)
690 X1=(B+C*A*D)/(C^2+D^2)
700 : OUTPUT R1,X1
710 RETURN
720 STOR: : SERIES TO PARALLEL CONVERSION
730 : INPUT RS,XS
740 RP=XS^2/RS+XS
750 XP=RS^2/XS+XS
760 : OUTPUT RP,XP
770 RETURN
780 ZTRANS: : TRANSFORMATION USING TRANSMISSION LINE
790 : INPUT Zo,ANG,Ro,Xo
800 IF ANG=90 THEN T=1.E100 ELSE T=TAN (ANG)
810 A=Zo*Ro @ B=Zo*Xo+Zo^2*T
820 C=Zo-Xo*T @ D=Ro*T
830 R1=(A+C*B*D)/(C^2+D^2)
840 X1=(B+C*A*D)/(C^2+D^2)
850 : OUTPUT R1,X1
860 RETURN
870 GAM: : CONVERTS Z TO GAMMA
880 : INPUT R,X
890 GAM=SQR ((R-Z)^2+X*X)/(R^2+X*X)
900 : OUTPUT GAM
910 RETURN
920 LOSS: : POWER LOSS DUE TO GAMMA
930 : INPUT GAM
940 ON ERROR GOTO 960
950 L=10*LGT (1-GAM^2) @ GOTO 970
960 OFF ERROR @ L=-1.E00
970 : OUTPUT LOSS (IN DB)
980 RETURN
990 END

```

Simplifying Propagation Loss Calculations

By Mel Bebee
Rockwell International

Engineers working with communication systems performance need to calculate propagation loss. A major component of this calculation is the free space loss between two isotropic antennas. Most engineers derive the formula each time it is needed by taking 10 log of the ratio between a spherical surface at a distance "D" ($4\pi D^2$) and the effective area of an isotropic radiator ($\lambda^2/4\pi$). Another formula for the above ratio is:

$$\text{Loss (dB)} = 20\log(F) + 20\log(D) + K \quad (1)$$

The problem arises in remembering the value of K. For F in MHz and D in meters, the value of K is -27.56 dB. Note that this value changes with various values of D. Table 1 shows the different values of K for the various distance units. Unless the engineer has this table handy or has memorized it, the equation above is useless.

A unit of distance can be defined for which K will equal 0. This distance is

300/(4 π) or 75/ π meters. The author has named this constant a "Perry," inspired by Beth Perry, a Rockwell engineer who conducted a propagation analysis. It is not uncommon to define a unit of distance for convenience in a particular field. Examples include fathoms, mils and fermis. The resulting equation is:

$$\text{Loss (dB)} = 20\log(F) + 20\log(D) \quad (2)$$

where F is in MHz and D is in Perrys.

Distance Unit (D)	K (in dB)
meters	-27.56
yards	-28.34
feet	-37.88
kilometers	+32.44
miles, statute	+36.57
miles, nautical	+37.79


Table 1. K values for various distance units.

Example 1

Find the free space loss (dB) at 10 MHz for a distance of 150 km between isotropic antennas.

$$\begin{aligned} \text{Distance} &= 150 \text{ km} \\ &= 150,000(\pi/75)\text{Perrys} = 2000\pi \\ \text{Loss(dB)} &= \\ &20 \log(10) + 20 \log(2000\pi) = 95.96 \text{ dB} \end{aligned}$$

Conclusion

Using this formula, the engineer only has to remember the conversion for Perrys (75/ π) and equation 2. This eliminates the problem of memorizing Table 1. 

About the Author

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Karl Loh, Jay Parton

1989 EDITORIAL CALENDAR

Issue	Shows & Special Coverage	Featured Technology	Industry Insight
JANUARY	<ul style="list-style-type: none"> • SMART IV (Surface Mount Expo) 	RF Power Amplifiers	Test Equipment
FEBRUARY	<ul style="list-style-type: none"> • RF Technology Expo 1989 • Aerospace Conference 	Small-Signal Amplifiers	GaAs Technology
MARCH		Filter Design	Power Transistors
APRIL	<ul style="list-style-type: none"> • National Association of Broadcasters (NAB) 	The RF Spectrum: I. Low and Medium Frequencies	Packaging Update
MAY	<ul style="list-style-type: none"> • Frequency Control Symposium • IEEE EMC Conference • Recruitment Issue/College Distribution 	The RF Spectrum: II. HF Technologies	Frequency Synthesis
JUNE	<ul style="list-style-type: none"> • MTT-S • European Microwave Show 	The RF Spectrum: III. VHF and UHF	Subsystems
JULY	<ul style="list-style-type: none"> • EMC Expo 	Electromagnetic Compatibility <ul style="list-style-type: none"> • 1989 Design Contest Results 	Filters
AUGUST	<ul style="list-style-type: none"> • Quartz Devices Conference • Antenna Measurement Techniques Association (AMTA) 	Crystal Oscillators and Filters	Attenuators and Switches
SEPTEMBER	<ul style="list-style-type: none"> • Coil Winding Show 	Test and Measurement Techniques	Inductors, magnetic materials (ferrite, iron powder)
DIRECTORY (SEPT.)	All Shows	DIRECTORY ISSUE Vendor Listings, Design Notes	
OCTOBER	<ul style="list-style-type: none"> • Old Crows • Ultrasonics Symposium • Electronica (Europe) 	System Design: Build-or-Buy Decision	SAW Update
NOVEMBER	<ul style="list-style-type: none"> • RF Expo East 1989 	Using Passive Components	RF Software
DECEMBER	Article Index	Mixers, Modulators and Demodulators	Cables & Connectors

PUBLISHER'S STATEMENT

FOR 6 MONTH PERIOD ENDING DECEMBER 1988

R. F. DESIGN

Cardiff Publishing Company
6300 South Syracuse Way
Suite 650
Englewood, CO 80111
(303) 220-0600

Official Publication of: None
Established: 1978
Issues Per Year: 12

No. NEC/12-88

No attempt has been made to rank the information contained in this report in order of importance, since BPA believes this is a judgement which must be made by the user of the report.

BUSINESS PUBLICATIONS AUDIT
OF CIRCULATION, INC.
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(212) 532-6880



FIELD SERVED

R.F. DESIGN serves engineers, managers, and other personnel responsible for the design and development of electronic systems and equipment using radio frequencies (RF), in the following industries: military, aerospace, ECM, EW, radar, land mobile, marine, navigation, CATV, broadcast, video and other communications; test and instrumentation; consumer components; electronics; and user industries. It is also distributed to labs, consultants.

DEFINITION OF RECIPIENT QUALIFICATION

Qualified recipients of R.F. DESIGN are engineering managers, engineers, members of technical staff, and other titled and non-titled personnel in the field served.

AVERAGE NON-QUALIFIED CIRCULATION 1. AVERAGE QUALIFIED CIRCULATION BREAKDOWN FOR PERIOD

	Copies		Qualified Non-Paid		Qualified Paid		Total Qualified	
			Copies	Percent	Copies	Percent	Copies	Percent
Advertiser and Agency	1,142	Single	35,474	13.2	102	0.3	35,606	100.0
Non-Qualified Paid	280	Group	-	-	-	-	-	-
Rotated or Occasional	-	Association	-	-	-	-	-	-
Samples	103	Bulk	-	0.3	3	-	3	-
All Other	1,563							
TOTAL	3,088	TOTALS	35,474	13.5	105	0.3	35,609	100.0

2. QUALIFIED CIRCULATION BY ISSUES WITH REMOVALS AND ADDITIONS FOR PERIOD

1988 Issue	Qualified Non-Paid	Qualified Paid	Total Qualified	Number Removed	Number Added	1988 Issue	Qualified Non-Paid	Qualified Paid	Total Qualified	Number Removed	Number Added
July			35,080	122	1,060	October			36,115	776	1,778
August			35,081	647	648	November			36,174	485	544
September			35,113	59	91	December			36,090	860	776
						TOTALS				2,949	4,897

3a. BUSINESS/OCCUPATIONAL BREAKDOWN OF QUALIFIED CIRCULATION FOR ISSUE OF NOVEMBER 1988

This issue is 1.9% or 678 copies above the average of the other 5 issues reported in Paragraph 2.

BUSINESS AND INDUSTRY	TOTAL QUALIFIED COPIES	PERCENT OF TOTAL	CLASSIFICATION BY TITLE				
			ENGINEERING MANAGER	ENGINEER, MEMBER TECHNICAL STAFF	Engineering Service Titles (Standards, Reliability, QC, Test Engineers)	President, Owner/Operator, Vice President, General Manager, Plant Manager	OTHER TITLED & NON-TITLED PERSONNEL & COMPANY COPIES
Government, Military, Aerospace, ECM, EW, Radar Manufacturers	1,134	3.1	193	722	193	15	11
a. EW, ECM and Radar Systems	5,185	14.3	1,103	3,362	581	103	36
b. Government Agencies other than Military	1,327	3.7	344	718	232	20	13
c. Military Communications Systems ...	4,573	12.6	937	2,915	576	122	23
d. Satellite and Space Systems	2,217	6.1	448	1,450	244	58	17
CATV & Broadcast Equipment	1,580	4.4	508	618	228	217	9
Land Mobile Equipment	2,120	5.9	529	994	241	338	18
Ground Communication Equipment Manufacturers	274	0.8	65	154	39	12	4
Video, Audio Equipment	648	1.8	168	322	79	74	5
Data Transmission/Computer Systems	2,167	6.0	505	1,167	313	173	9
Instruments & Test Equipment	3,466	9.6	732	1,844	537	309	44
Medical Electronics Equipment	632	1.7	143	365	61	59	4
Industrial Controls & Power Supplies ...	689	1.9	168	329	84	103	5
Consumer Electronics Equipment	1,168	3.2	267	545	182	171	3
Components & Subsystems	3,673	10.2	878	1,804	393	535	63
Aviation, Marine, Navigation Systems ..	2,438	6.7	497	1,497	353	77	14
Lab or Consultant	2,299	6.4	479	968	241	577	34
Users of Electronic Equipment	406	1.1	108	158	99	32	9
Others Allied to the Field	178	0.5	41	88	19	16	14
TOTAL	36,174	100.0	8,113	20,020	4,695	3,011	335
PERCENT	100.0	-	22.4	55.4	13.0	8.3	0.9

3b. QUALIFICATION SOURCE BREAKDOWN FOR ISSUE OF NOVEMBER 1988

QUALIFICATION SOURCE	Qualified Within			Qualified Non-Paid	Qualified Paid	Total Qualified	Percent
	1 year	2 years	3 years				
I. Personal direct request from the recipient:							
a. Written	33,323	2,851	-			36,174	100.0
b. Telecommunication	-	-	-			-	-
II. Request from recipient's company:							
a. Written	-	-	-			-	-
b. Telecommunication	-	-	-			-	-
III. Association or society membership:							
a. Individual	-	-	-			-	-
b. Organizational	-	-	-			-	-
IV. Communication from recipient or recipient's company (other than request):							
a. Written	-	-	-			-	-
b. Telecommunication	-	-	-			-	-
V. Sources other than above (listed alphabetically):							
Association rosters and directories	-	-	-			-	-
Business directories	-	-	-			-	-
Independent field reports	-	-	-			-	-
Licensees-Federal, State or Local Government	-	-	-			-	-
Manufacturer's, distributor's and wholesaler's lists	-	-	-			-	-
Other sources	-	-	-			-	-
TOTALS	33,323	2,851	-			36,174	100.0
Percents	92.1	7.9	-			100.0	

3c. MAILING ADDRESS BREAKDOWN OF QUALIFIED CIRCULATION FOR ISSUE OF NOVEMBER 1988

	Qualified Non-Paid	Qualified Paid	Total Qualified	Percent
Individuals by name and title and/or function			36,137	99.9
Individuals by name only			18	0.1
Titles or functions only			3	-
Company names only			13	-
Bulk copies			3	-
TOTALS			36,174	100.0

4. GEOGRAPHICAL BREAKDOWN OF QUALIFIED CIRCULATION FOR ISSUE OF NOVEMBER 1988

State & Zip Code	Qualified Non-Paid	Qualified Paid	Total Qualified	Percent
039-049 ME			36	
030-038 NH			396	
050-059 VT			28	
010-027 MA			2,231	
028-029 RI			84	
060-069 CT			494	
NEW ENGLAND			3,269	9.0
100-149 NY			2,539	
070-089 NJ			1,574	
150-196 PA			1,124	
MIDDLE ATLANTIC			5,237	14.5
430-458 OH			661	
460-479 IN			728	
600-629 IL			1,461	
480-499 MI			379	
530-549 WI			222	
EAST NO. CENTRAL			3,451	9.5
550-567 MN			410	
500-528 IA			337	
630-658 MO			334	
580-588 ND			16	
570-577 SD			26	
680-693 NE			64	
660-679 KS			219	
WEST NO. CENTRAL			1,406	3.9
197-199 DE			23	
206-219 MD			1,476	
200-205 DC			189	
220-246 VA			967	
247-268 WV			22	
270-289 NC			270	
290-299 SC			59	
300-319 GA			435	
320-349 FL			1,496	
SOUTH ATLANTIC			4,937	13.7
400-427 KY			45	
370-385 TN			172	
350-369 AL			264	
386-397 MS			37	
EAST SO. CENTRAL			518	1.4
716-729 AR			35	
700-714 LA			44	
730-749 OK			106	
750-799, 885 TX			1,707	
WEST SO. CENTRAL			1,892	5.2
590-599 MT			24	
832-838 ID			38	
820-831 WY			15	
800-816 CO			690	
870-884 NM			310	
850-865 AZ			1,038	
840-847 UT			237	
889, 890-898 NV			121	
MOUNTAIN			2,473	6.8
995-999 AK			14	
980-994 WA			599	
970-979 OR			300	
900-961 CA			8,754	
967-968 HI			42	
PACIFIC			9,709	26.9
UNITED STATES			32,892	90.9
969 & 006-009 U.S. Territories			41	
Canada			886	
Foreign			2,276	
APO/FPO			79	
TOTALS			36,174	100.0

QUALIFIED PAID CIRCULATION

5. PRICES

6. LENGTH OF SUBSCRIPTIONS

7. SOURCES

PARAGRAPHS 5 THROUGH 10 ARE NOT APPLICABLE

8. PREMIUMS

9. STATUS OF QUALIFIED PAID SUBSCRIPTION PAYMENTS

10. RENEWAL PERCENTAGE OF QUALIFIED PAID CIRCULATION

11. ADDITIONAL DATA

PUBLISHER'S AFFIDAVIT

We hereby make oath and say that all data set forth in this statement are true.

Robert Searle

President

Patricia Shapiro

Circulation Director

(At least one of the above signatures must be that of an officer of the publishing company or its authorized representative.)

Date signed

December 13, 1988

State

Colorado

County

Arapahoe

IMPORTANT NOTE:

This unaudited publisher's statement has been checked against the previous audit report. It will be included in the annual audit made by Business Publications Audit of Circulation, Inc.

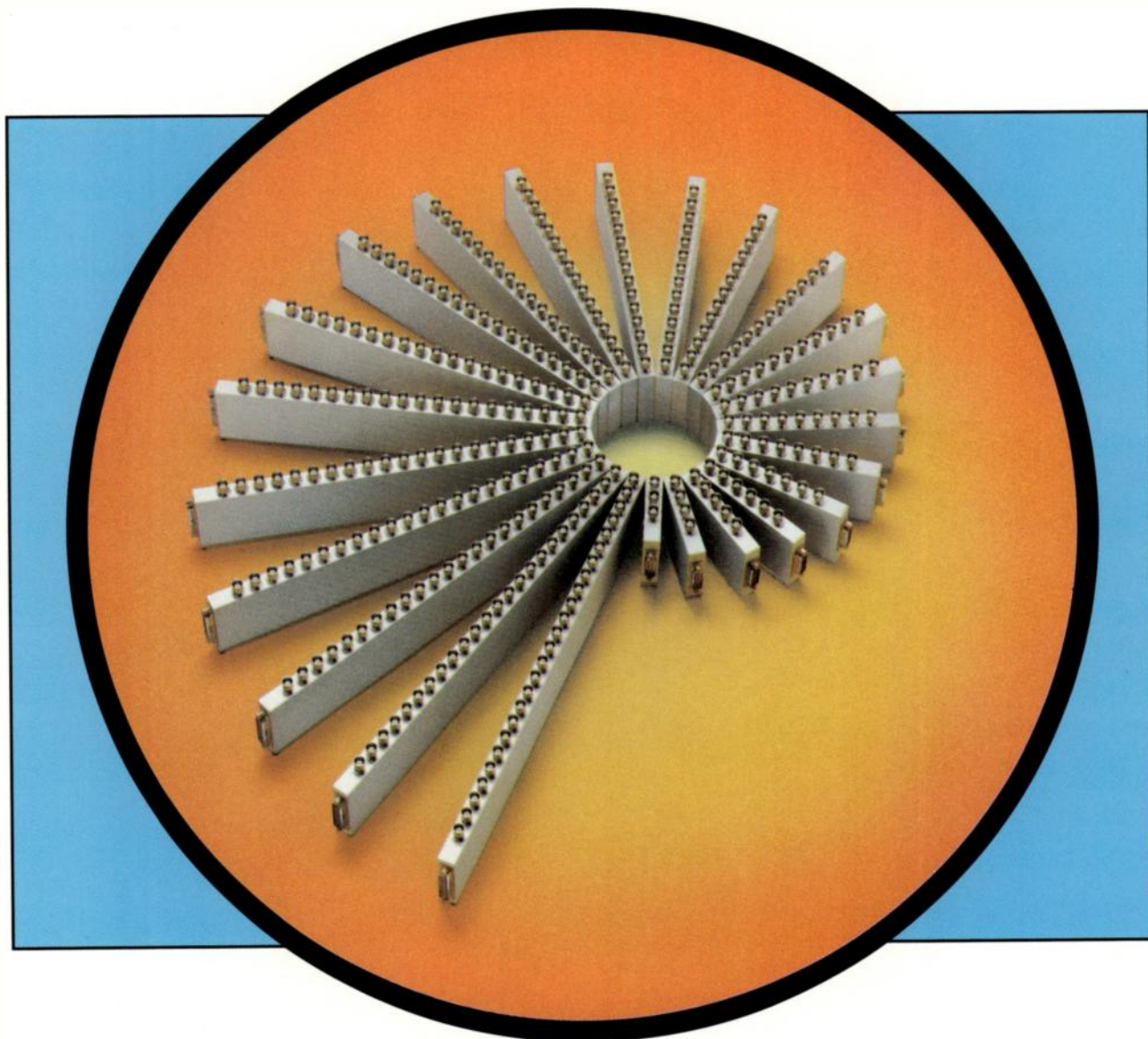
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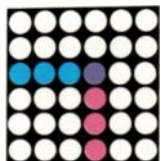
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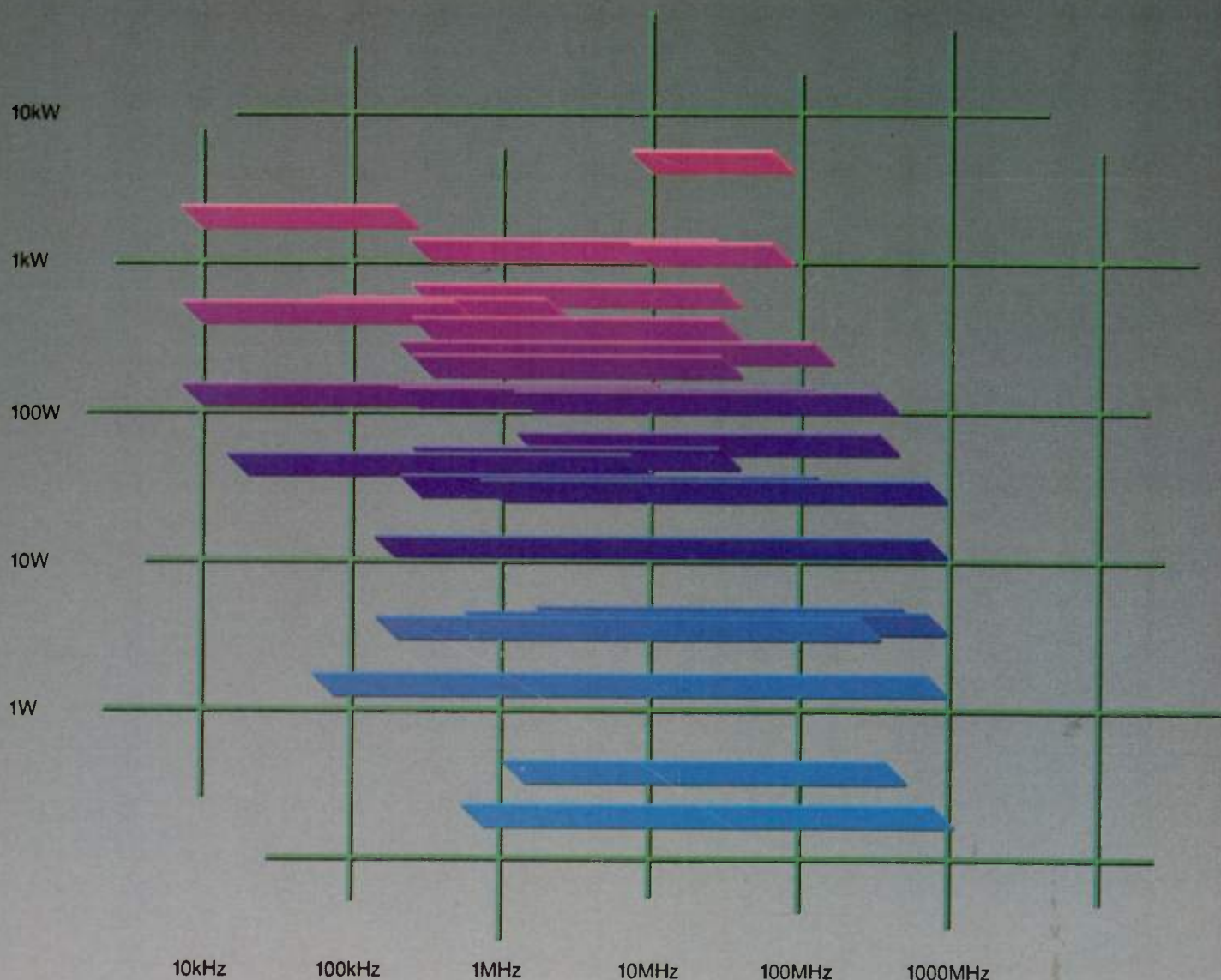
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