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# Good News

#### SNELL

The Type D three-way loudspeaker has flat response to within ±1.75dB from 36-20kHz (halfspace anechoic), both on- and as much as 25° off-axis. The design supports use with amplifiers rated from 20-200W/channel. Driver configuration includes a 1" titanium dome tweeter with a pleated titanium surround, a rear-firing supertweeter, a 5" midrange driver, and an 8" woofer. Snell Acoustics, 143 Essex St., Haverhill, MA 01832, (508) 373-6114, FAX (508) 373-6172. Reader Service #70



#### O POLYDAX

## WILMSLOW AUDIO

The ATCK50 and ATCK100 kits offer highly damped hi-fi/studio drive units, soft-diaphragm midrange units, high-frequency units (damped fabric diaphragm), and edge-wound ribbon voice coils. Optimized suspension systems allow maximum cone extension. The kits represent a collaboration between Wilmslow Audio and ATC Loudspeaker Technology. Wilmslow Audio Ltd., Wellington Close, Parkgate Trading Estate, Knutsford, Cheshire WA16 8DX, England, (0565) 650605, FAX (0565) 650080.

Reader Service #59

The AW025S1 tweeter combines a pure titanium diaphragm with a soft polymer suspension. Its moving assembly integrates a high-energy acoustic lens with an adjusted suspension. The motor structure is magnetically shielded for audio/video applications. Frequency range is 3-20kHz, with a free-air resonance of 1.5kHz, power handling of 80W RMS, and a sensitivity rating of 92dB 1W/1M. Polydax Speaker Corp., 10 Upton Dr., Wilmington, MA 01887, (508) 658-0700, FAX (508) 658-0703. Reader Service #52

# PULSE CAL OUT 1 PROJES - 2

#### **O LIBERTY**

The IMP Audio Analyzer is a Fast Fourier Transform spectrum, impedance, and network analyzer for use with IBM-compatible computers. The device, which makes frequency response and phase measurements, connects externally to the computer via the standard printer port. The IMP's graphically based software enables extraction of quasi-anechoic acoustic measurements, transient response analysis in 3-D "waterfall" format, and correction for microphone response, among other features. Liberty Instruments, Inc., PO Box 1454, West Chester, OH 45071. Also available from Old Colony Sound Lab, PO Box 243, Peterborough, NH 03458-0243, (603) 924-6371, FAX (603) 924-9467. *Reader Service #54* 



#### C AUDIOCONTROL

The new C-101 Series III octave equalizer has a built-in digital pinknoise test generator and a realtime audio spectrum analyzer. Octave-spaced sound controls allow for adjustment while pairing left and right sliders. The Series III can be connected to any home stereo, comes with a calibrated microphone, and features an 18dB/octave Chebychev alignment subsonic filter. AudioControl. 22313 70th Ave. West, Mountlake Terrace, WA 98043, (206) 775-8461, FAX (206) 778-3166. Reader Service #57

#### C&S AUDIO

The Woofer Tester is a combination hardware and software system that runs on any IBM-compatible PC with EGA or better graphics capability and at least one RS-232 serial port. The software (one 3.5" diskette) will give all measurements needed to characterize a raw driver and design and build closed- or vented-box woofer systems, including woofer and system resonant frequencies, Q, and system tuning ratio. Test results are available both onscreen and written to the PC disk drive, C&S Audio Labs, Floyd, VA, (301) 498-8737.

Reader Service #53

# Good News

### POLYDAX

The latest line of woven Kevlar<sup>®</sup> cone mid-bass speakers includes the HT100K0 (4"), HT130K0 (5¼"), HT170K0 (6½"), and HT210K0 (8"). Design features include high-loss rubber surrounds, large (20 oz.) magnet structures, and high-temperature voice coils wound on aluminum voice coils. Polydax Speaker Corp., 10 Upton Dr., Wilmington, MA 01887, (508) 658-0700, FAX (508) 658-0703. *Reader Service #51* 



## **O** NEUTRIK

The A2 Audio Measurement System controls and maintains transmission and reproduction sound quality. The stand-alone unit can be used with or without computer interface to proof and service broadcasting and recording equipment. Measurement functions include harmonic and intermodulation distortion, noise, phase shift, and crosstalk. The device has a built-in digital storage audio oscilloscope and spectral analyzer. For more information contact Vincent DeSouza, Neutrik Instruments, Inc., (514) 344-5220.

Reader Service #55

## WORLD COLLEGE The Cleveland Institute of Electron-

ics has announced the organization of its affiliate college, offering bachelor's degrees in electronics engineering technology (pending approval from the Virginia Council of Higher Education). The college offers extensive electronics and related courses, and is accredited for home study. World College, 1776 E. 17th St., Cleveland, OH 44114, (800) 243-6446. *Reader Service #75* 

## NORTH CREEK

Three new Unlimited Series kits from North Creek (for crossover, cabinet, and Okara loudspeakers) all feature the Scan-Speak D2905 fabric dome tweeter, Vifa and Scan-Speak woofers. Unlimited Crossovers are housed in their own dedicated enclosure, and feature 10 AWG inductors, polypropylene caps, and Ohmite 1% precision power resistors. All components are matched to  $\pm$ 1%. Also new is the complete passive crossover replacement for the B&W 802 Series II and III loud-speaker systems. North Creek Music Systems, Route 8, PO Box 500, Speculator, NY 12164, (518) 548-3623.

Reader Service #58

### ⊃ PYLE INDUSTRIES

Pyle NeoDome™ tweeters utilize focused field neodymium magnet structures and chemically laminated titanium domes. Other features include ferrofluid-cooled motors and computer-designed housings. Three models are available with surface, flush or pivotmount housings. Pyle Industries, Inc., 501 Center St., PO Box 620, Huntington, IN 46750, (219) 356-1200, FAX (219) 356-2830. *Reader Service #62* 

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The second generation of the HCA-1200 power amp has been certified by Lucasfilm THX™ Home Cinema. It incorporates new John Curl circuitry and 5-mm-thick, 19" rack-mount panel. The HCA-1200<sup>II</sup> has output of 205W/channel into  $8\Omega$ , 2 × 315W into  $4\Omega$ , and >630W bridged into 8Ω. A 1kVA toroid transformer with 60,000µF power supply sustains continuous 40A and peak 57A current with 2dB headroom. Curl's design includes ten pairs of 60MHz, 15A bipolar output transistors. Parasound Products, 950 Battery St., San Francisco, CA 94111, (800) 822-8802.

Reader Service #65

Continued on page 6



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#### Continued from page 4

#### IMPULSE

The H6 hom loudspeaker incorporates the same high-frequency unit used in the H1 and H2 models with compact design. Cabinet dimensions are  $36^{\circ} \times 7.5^{\circ} \times 14^{\circ}$ . Nominal impedance is  $8\Omega$  with sensitivity of 89dB (2.83V at 1M). Low and midfrequencies are produced by a doped paper coned drive unit which is backloaded by a horn. Impulse Loudspeakers, 5 High Parade, Streatham High Road, London SW16 1EX, (081) 769-5726, FAX (081) 769-0353. **Reader Service #64** 

#### M&K

The S-90 speaker system was designed to achieve a timbre-match with other front-channel speakers to avoid the sonic discontinuity which occurs when sound pans across the left, center, and right channels of unmatched speakers. Miller & Kreisel Sound Corp., 10391 Jefferson Blvd., Culver City, CA 90232, (310) 204-2854, FAX (310) 202-8782. *Reader Service #74* 

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#### O DALANCO SPRY

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Good News

The Model 310A DSP and Data Acquisition Board was designed for the IBM PC/AT and bus-compatible microcomputers. Based upon the Texas Instruments TMS320C31, the floating-point DSP operates at 33MHz for up to 33 MFLOP performance. The Model 310A provides data acquisition for four differential channels at 14-bit resolution, with programmable gain and a maximum sampling rate of 150kHz. Dalanco Spry, 89 Westland Ave., Rochester, NY 14618, (716) 473-3610, FAX (716) 271-8380.

Reader Service #63

#### M&P TECHNOLOGIES

All Selenium products (woofers, horn drivers, horns and tweeters) are now available from M&P. The Pro Unit Woofers (12", 15", 18") feature 1kW continuous power handling with up to 100dB SPL (1W/1M). Factory computeraligned set screws allow for replacement of cone, spider, former and voice coil. Selenium also manufactures customized speakers. M&P Technologies, Inc., 75 E. Uwchlan Ave. #128, Exton, PA 19341, (800) 355-0500, FAX (215) 524-5531.

Reader Service #67

#### O SYON

TRU-BOND<sup>®</sup> premixed, two-part frozen epoxy compounds are available in a variety of standard and custom formulations, including electrically and thermally conductive, and copper- and silver-filled. Packaged in 2–60 cc syringe kits, the compounds cure to a 85 Shore D hardness, with flexural strength of 6,000 psi. They are resistant to moisture, acids, and alkalis. Syon Corp., 280 Eliot St., Ashland, MA 01721, (508) 881-8852, FAX (508) 881-4703.

Reader Service #66

#### **O DECADE ENGINEERING**

LoudMan<sup>™</sup> is a 30W/channel stereo power amplifier kit designed for direct connection to the earphone jack of a Walkman<sup>™</sup> or other low-power stereo product. It can also drive high-performance 8Ω speakers from portable stereos. Kit includes glass-epoxy PC board, heatsink, all electronic components, assembly instructions, and Theory of Operation document. Total power is 60W and load impedance  $\ge 8\Omega$ , with typical midband distortion below 0.05%. Decade Engineering, 2302 5th St. NE, Salem, OR 97303, (503) 363-5143, FAX (503) 399-9747. **Reader Service #61** 



# Good News



#### **O PC INSTRUMENTS**

The 420 and 430 Series oscilloscope boards for PCs provide two independent channels in one expansion slot. The single-channel scopes can be upgraded to dual channel. Both are available with two different attenuators, and provide 200MHz bandwidth, 500 ps/div minimum timebase setting, and 200 gigasample/sec. equivalent sampling rate. Optional Bench-Com<sup>™</sup> software handles scope controls and waveform display, and allows communication with existing software. PC Instruments, Inc., 9261 Ravenna Rd., Bldg. B11, Twinsburg, OH 44087, (216) 487-0220, FAX (216) 425-1590. Reader Service #71

#### SCANTEK

The new brochure on the precision Type 116 Sound Level Meter from Norsonic describes features and specifications. With an 80dB dynamic range, the meter simultaneously measures peak and RMS values of A- and C-weighted noise levels. Data is presented both numerically and graphically, and the unit is PC compatible. Scantek, Inc., 916 Gist Ave., Silver Spring, MD 20910, (301) 495-7738, FAX (301) 495-7739.

**Reader Service #68** 

### POLYDAX

AUDIO TALK, Polydax's new quarterly newsletter, informs speaker enthusiasts of new developments, products, and ideas within the industry. Articles are written by professionals both inside and outside the company. Topics covered in the first edition include break-

#### MEADOWLARK

Keldamp is an adhesive-backed gasket material made from a highdensity compound with damping characteristics which control the transfer of energy between driver and enclosure. The material does not flatten out, but retains its resil-

#### ■ B&W

The Matrix 803 Series 2 loudspeaker features three 6" drivers. and a sloping shelf: cabinet dimensions have also been reduced. Frequency range is 20Hz-22kHz; sensitivity is 90dB SPL (2.83V/1M); nominal impedance is 8Ω. B&W Loudspeakers of America, PO Box 653, Buffalo, NY 14240, (800) 387-5127. Reader Service #72

iency to ensure an air-tight seal. Keldamp is available in 1/16" ×  $\frac{1}{2}$ " × 108' rolls, and in other thicknesses and widths. Meadowlark Audio, 1648 Marbella Dr., Vista, CA 92083, (619) 598-3763. Reader Service #60

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#### DESIGN ACOUSTICS

Features of the 15-inch-high DA800 bookshelf model speaker include broad bandwidth (150Hz-4kHz), 8" downward-firing woofer, and small front baffle. Diffraction effects are further reduced with Sonofoam acoustic foam. Design Acoustics Division, A.T.U.S., Inc., 1225 Commerce Dr., Stow, OH 44224, (216) 686-2600.

Reader Service #56

#### TRI-STATE

The new catalog of home stereo, automotive, and professional speakers, parts and accessories, refoaming and reconing services is now available. Normally a \$1 value, the catalog is currently being offered free to any speaker builder who calls. Tri-State Loudspeaker, 650 Franklin Ave., Aliquippa, PA 15001, (412) 375-9203.

Reader Service #69

throughs in loudspeaker technology and new products entering the market. Polydax Speaker Corp., 10 Upton Dr., Wilmington, MA 01887, (508) 658-0700, FAX (508) 658-0703.

Reader Service #73





**The Art and Science...** of loudspeaker system development today has become more complex than ever before. Competition is tough, and to compete each design must perform to the best of its ability, and make the most out of every dollar's worth of transducer cost. The simple approach of choosing a combination of seemingly appropriate transducers coupled with ordinary networks and filters, has given way to a painstaking process of meticulously blending selected transducers in combination with carefully devised and matched crossover designs.



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The upcoming 1994 year marks two anniversaries: This magazine turns 15, and *Audio Amateur* will celebrate a quartercentury of publication. I confess to being somewhat surprised by this turn of events, but, on the whole, rather pleased to have seen a pair of brainchildren survive for this time span.

Much is new with your publishing company. We are now a corporation for any number of reasons—which mostly escape me most of the time. The folks we pay to advise us about such things all seem to think it is a good idea. So we are. For legal reasons I sign things these days with a pretentious pair of titles, but my favorite appelation still continues to be: Editor/Publisher.

A number of new people have joined our ranks recently. Although Laurel Humphrey is not new to the company, having served as a circulation clerk for a year previously, she returned to us as Circulation and Promotion manager last June. Already activities in her important department have become greener since a national brokerage now handles our mailing lists and we have added several new, large newsstand outlets, one of them being Tower Magazines, a division of Tower Records. She also manages circulation promotion and a pesky computer network we are trying to subdue to our will.

Martha Povey has joined us as National Advertising Director, coming to us from stints at the WGE and IDG organizations here in Peterborough. Her work is being aided considerably by those we recently asked to help us by responding to a survey form about who you are, what you do, how you use this magazine and just how you practice your hobby of speaker building. We'll have a full report in the magazine early next year. Meanwhile Martha is getting to know our advertisers and working with them to bring you details about their offerings of wares and services.

And speaking of news, we are exceptionally pleased to announce an additional two issues of *Speaker Builder* beginning in 1994. This brings the total to eight per year, giving us room for more articles, pictures, reports and reviews. That means a price increase also, unfortunately, but we still welcome your renewal at the 1993 rate for six issues if you act before December 31 of this year. The new rate of \$32 per year (U.S. only) becomes effective as of January 1, 1994.

You may have noticed a new look in our other periodicals having to do with typefaces and page design. The "look" now comes to *Speaker Builder* as well. By now everyone has heard of desktop publishing. While we are not, as yet, fully "desktop" we are moving closer by the week. In time, our pages will go to our printer on a removable data storage disk or on a high speed modem, rather than a large package of film negatives from which printing plates are made.

Text has, for many years, been computer generated and fed to a digital typesetting machine. But the latter used photosensitive paper which required development in a large machine full of water and chemicals. As this issue goes to press, the "old" typesetter leaves the building bound for a nearby printing establishment to upgrade a system even further behind technologically than ours. You will see further changes in the look of our pages as the months pass. We welcome your comments, as always.

You might be interested to know that our overseas subscriber lists are growing faster, in some cases, than those within the U.S. Almost one out of four of *Glass Audio*'s subscribers lives outside the United States. This reach to the other countries of the world includes the books this company publishes through its Old Colony division. Vance Dickason's bestselling *Loudspeaker Design Cookbook* is now also available in German and Portuguese translations. We hope to announce other versions soon.

Speaking of new books, we are offering a couple of new ones full of old ideas. *Audio Anthology* No. 5 is now off the press. This popular series is a resurrection of one published by the precursor of *Audio* magazine in the early sixties. The other oldie resurrects a selection of editorials from *Audio Amateur* over its first fifteen years and puts them between hard covers. It offers a few predictions, meditations on music and other matters under the title *Of Mockingbirds and Other Irrelevancies*. Pick up a copy at your local 7-11 or wherever popular titles are sold.

I find it interesting to note how far we have moved in this quarter century span. Life for these periodicals began in 1969 in the attic room of 307 Dickinson Avenue, Swarthmore, Pennsylvania. We did our work on blue gridded sheets of legal size paper, typing on a used IBM Selectric typewriter. For headlines we reached for presstype, and changed the Selectric's golfballs to do italic or boldface.

As I sit in our building in Peterborough twenty-four years later, I am capturing this text in Times New Roman, 12 point (and two keystrokes enables italic) under the guidance of Microsoft Word for Windows on a 486/50MHz IBM clone, with Vaughan Williams' Symphony No. 4, conducted by Sir Adrian Boult, coming to my ears through headphones being fed by a CD ROM player in the top slot of the computer. The distance somehow seems a whole lot more than 24 years. But *what* a journey. I would not have missed it for anything.—E.T.D.





#### The Staff

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Speaker Builder is published bi-monthly in the interest of the art and craft of speaker building.

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## About This Issue

The articles appearing in this, the final issue of 1993, are testimony to the ingenuity of speaker builders. Each design bears the indelible stamp of its creator, as evident in our lead article. Already well-known for his show horns, Contributing Editor **Bruce Edgar** describes the process of designing, constructing, and testing "The Monolith" (p. 12). We think you'll be impressed by this massive undertaking.

Do your speakers take up too much space? As a solution, **Bill Fitzmaurice** proposes "Orbiting Satellites," beginning on page 26. It just goes to show what a little furniture remodeling can do for your creativity. In this case, it transformed Bill's beloved Klipsch clones into something out of this world.

**Ralph Gonzalez** brings us back to terra firma with variable-Q formulas in "Real-World Three-Way Crossovers" (p. 30). Whether your crossovers are ideal or the nonideal variety, the computer program Ralph describes will enable you to choose exactly the right Q for your system.

Riddle: How is a loudspeaker like a string instrument? As Ernie Pfannenschmidt explains, it's all in the enclosure. His design begins with a simple analogy—in fact, he calls it "The Simplex—and, for ease of construction, it lives up to its name. Simple is definitely better, as you'll learn beginning on page 32.

Faithful SB readers are familiar by now with **Bill Waslo's** IMP: it was born in our pages almost one year ago. This chapter in the IMP saga describes noise stimulus, maximum length sequence test signal, and the new version M software with all its bells and whistles, as "The IMP Goes MLS" (p. 40).

For an idea of how an "average" speaker builder appears to the uninitiated, turn to page 38, where **Nancy MacArthur** describes "Living With a Speaker Builder." Obviously, she's totally objective and unbiased! The tables have turned with this gritty exposé of the "other side."

Rounding out this issue, **Bob Wayland** sets the record straight in "Wayland's Wood World" (p. 50); the resources of several *SB* readers have been pooled in "Tools, Tips & Techniques" (p. 56); and **Stephen Katz** describes his gym horn in "Craftsman's Corner" (p. 60).

See you next year!







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# THE MONOLITH HORN

By Bruce C. Edgar Contributing Editor

A fter my "Show Horn" article appeared (SB 2/90, p. 10), I received a number of inquiries about designing a bass horn that would go down to 30-40 Hz. One SB reader, Fred Ireson, requested a 40Hz horn for a 15'' driver, and I have been refining the design and construction details ever since.

The 40Hz horn is shown in *Photo 1*. Since the structure is rather imposing, I have nicknamed it the "Monolith." Despite some unusual features, such as a mouth that exhausts out of the bottom and a top-mounting driver compartment, the modular construction is well within the capabilities of a skilled home craftsman with a table saw.

### **EXPERIMENTAL STAGE**

Experimentation helped me avoid potential problems. Since bass horns can be huge, designers try tricks such as 180° folds, reducing the mouth size, and shortening the length in an effort to keep the overall volume to reasonable proportions. Unfortunately, if they are not applied intelligently, these techniques can lead to numerous response anomalies.



PHOTO 1: The Monolith Horn.

One of the classic horn-folding designs utilized a series of  $180^{\circ}$  folds that approximated an exponential flare in steps.<sup>1</sup> The overall response of the rear-loaded bass horn (*Fig. 1*) rolls off above 200Hz, and it is unclear whether this is due to the cone mass or the folds. If we move the rolloff up to 400Hz, however, we can achieve good wide-band performance with a reduction in volume.

Before conceiving the "Show Horn," I was asked by a reader to design a 50Hz corner horn with the smallest volume (for shipping overseas) plus a wide bandwidth to mate with a 500Hz midrange horn. In my naïveté, I set off on my mission not realizing the potential conflicts in the design requirements. To attain the 500Hz bandwidth, I chose to use the EVM 12L driver, which has a mass rolloff over 500Hz. *Figure 2* is a design sketch featuring several 180° folds with a top-mounted driver and a bottom exhaust mouth—precursor to the Monolith.

l proceeded to build the bass horn and measure its response (*Fig. 3*). You can see the big 20dB "hole" between 300 and 400Hz, which is clearly unacceptable. At first, I didn't have a clue as to the root cause, but after some discussion with Dave Rowe and other colleagues, we arrived at the concept of placing the corner reflectors along the diagonal. When I replaced the existing corner reflectors with larger ones, the response hole partially filled up (*Fig. 4*), indicating that this approach was leading in the right direction.

BACK







dropout between 300–400Hz.

After the Show Horn article appeared, several people questioned why a diagonal reflector would make such a difference over a radius bend (*SB* 2/90, *Figs.* 4 and 5, p. 14). The radius bend dimensions are still a fraction of a wavelength (45'' at 300Hz), which satisfies Olson's conditions for proper horn folds.<sup>2</sup>

If you look at a 90° bend from a shortwavelength perspective, a reflector along the diagonal makes more sense. *Figures 5* and 6 graphically show Huygen's construction principle of wave fronts traversing a 90° bend using both types of reflectors. As you can see from *Fig. 5*, a radius-bend reflector gives both backward- and forward-traveling waves in response to the initial wave fronts, and those coming out of the bend are incomplete. Com-



FIGURE 5: Huygen's wavefront construction for a 90° duct bend with a radius reflector. "I" is the incident wave; "R1" and "R2" are the reflected waves.



FIGURE 6: Huygen's wavefront construction for a 90° duct bend with a diagonal reflector. "I" is the incident wave; "R" is the reflected wave.



FIGURE 4: Response of 1984 design with diagonal reflectors.

pare that with *Fig.* 6, where no backward-traveling wave exists, and the forward-traveling wave fronts coming out of the bend are completely reconstructed.

A 180° bend can be modeled as an acoustical inductance, and a horn with many folds can be modeled as a series of acoustical transmission-line strips separated by inductances (*Fig.* 7).<sup>3</sup> At the frequency where the length of the transmission-line strip is half a wavelength (at 300Hz, 22.5"), any transmission line will transfer to the input the impedance seen at the load end.

If a series of transmission-line strips of equal length are separated by inductances, the input impedance at the half-wavelength condition becomes a bunch of inductances in series. In the case of a multifolded horn where the bends are separated by equal lengths, the inductance load condition at the half-wavelength will swamp the mouth impedance and the response is choked off at that frequency.

In mathematical terms, the frequency for the half-wavelength condition is:

$$f = \frac{c}{2 \times 1}$$

where:

c = the speed of sound

l = the length of the horn section between bends

In the case of Fig. 2, the length between folds



PHOTO 2: Throat partitions mounted to the top.

was 191/2'', which corresponds to a half-wavelength frequency of 345Hz—right in the response hole of *Fig. 3*. In the case of Olson's horn example, l = 21'', for a null frequency of 321 Hz. The bass-horn response has a very sharp rolloff at 300Hz, which leads me to speculate that his horn was indeed affected by the null caused by reflections at the bends and the equidistant spacing between bends.

I discovered during my investigation of horn bends that the same principle is used to design mufflers. A muffler is a series of pipe lengths separated by small volumes which are acoustic capacitances. By adjusting the pipe lengths and volumes, you can design a very effective acoustical stop-band filter. So between 300 and 400Hz, my original horn design was behaving like a muffler!

A diagonal reflector will help, but not completely cure, the response ills occurring with 180° bends. You should first try to reduce their number or make them less severe. You can also use the nulling phenomenon from the bends to shape the horn's upper-frequency cutoff to your advantage, such as in a subwoofer horn.

#### **DESIGN STAGE**

After some discussion with Fred Ireson, we settled on the JBL 2220H 15" pro driver. Using formulas from the Show Horn article, its T/S parameters ( $f_S = 37$ Hz,  $Q_{ES} = 0.18$ , and  $V_{AS} = 10.5$  ft.<sup>3</sup>) give an optimum throat size of 56 in.<sup>2</sup> and a mass rolloff of 411Hz. Even though a











FIGURE 9: Side view of the Monolith bass horn.

resonance frequency of 37Hz would allow a lower flare frequency, I chose 40Hz to keep the size to manageable proportions. For the best response down to the flare cutoff frequency, I selected a hyperbolic exponential expansion of M = 0.6. Even with a one-eighth-sized horn, the path length is over 7' and the mouth size is nearly 8 ft.<sup>2</sup> (1,133 in.<sup>2</sup>), which gives you an idea of its imposing size.

The concept of a bottom exhaust for the mouth is not new: both the Lowther TP-1 and the Gately Super Horn used it.<sup>4,5</sup> With a bottom exhaust, the mouth can be wrapped around the horn base perimeter. If the two side and front widths add up to a length Lm, then



FIGURE 11: Sawing guide for 3/4" plywood.

the height (h) of the exhaust opening above the floor is h (in.) = 1,133/Lm.

In the Monolith's case, this distance from the floor turned out to be 13" for a wall position (two

TABLE 1				
MONOLITH H	ORN PARTS LIST			
PART	DIMENSIONS			
A	6" × 143∕₄"			
В	6" × 237⁄8"			
С	6" × 8₅⁄8"			
D	81∕2" × 27"			
E	63⁄4" × 34"			
F	151⁄2" × 34"			
G	16″ × 34″			
H (2)	101⁄2" × <b>34"</b>			
	71⁄4" × 34"			
J	6" × 157⁄8"			
Sides (2)	241⁄2" × 27"			
Тор	241⁄2" × 34"			
Front, back (2)	34" × 261⁄4"			
Back chamber side 1 (2)	123∕₄" × 171∕₂"			
Back chamber side 2 (2)	123⁄4″ × 16″			
Back chamber top	17V <sub>2</sub> " × 17V <sub>2</sub> "			

side areas and a front area forming a mouth) and about 18" for a corner position (one side area and a front area); however, this separation distance can be adjusted for the smoothest response, as we will examine later. The mouth's close proximity to the floor allows for good coupling to the acoustical images below the floor and behind the wall. The formation of these images multiplies the effective mouth area to provide good bass.

#### MONOLITH CONSTRUCTION

The Monolith's throat manifold layout and side view are shown in *Figs. 8* and 9, respectively. The former is similar to the Show Horn; however, with the back chamber on top, you can locate the throat and driver in an optimum position and not have to leave room for the back side duct. With a 15" driver, a top mounting is a decided advantage in reducing the horn's depth. The internal depth *Continued on page 16* 

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QTY 5,000	10ohm 5 watt 5%	DESCRIPTION Wire wound Resistor, Sand cast	3/8" Sq x 7/8" long, 1-1/2" leads, axial	PRICE 10 for \$1.50
21,000	4.7mfd Mylar HitachiCap.	100V, 1/4" x 1/2" oval by 1-3/16" long	1/2" long solid leads, axial	10 for \$4.00
34,000	2.2mfd Mylar Capacitor	100V, 10%, 3/8"D x 1" long	Yellow, 1 - 5/8" long solid leads, axial	10 for \$2.00
41,000	5mfd Mylar Elpac Capacitor 4.6mfd Mylar West-Cap Capacitor; -	50V, 10%, 1/4" x 3/8" oval by 1-1/4" long	2-1/8" long solid leads, axial 2-1/4" long solid leads, axial	10 for \$4.00 10 for \$9.00
399	60mfd Solen Polypropylene Capacito		2" long solid leads, axial	\$10.00 cach
41	47mfd Solen Polypropylene Capacito	*	2" long solid leads, axial	\$8.00 cach
150	45mfd Solen Polypropylene Capacito		1-3/4" long solid leads, axial	\$8.00 each
	25% discount on all Carli Mylar Cap	acitors: 1, 2.7, 3.3, 3.9, 4.7, 6, 6.8, 8, 10, 12, 15, 20 mfd.	Cannot be used with other discounts; does not apply to dealer and	OEM sales.)
400	.15mH Iron core Inductor; 24awg wi		1/2" D x 2" long	10 for \$5.00
2950	.2mH Air core Inductor; 19awg wire		1-1/2" D x 1-1/8" tall	10 for \$7.50
1,400	2.0mH Air core Inductor; 19awg wind		2" D x 1" tall	\$1.75 each
650			I grill with mounting ring. (Same restrictions as above for dealer C, 96.5dB, 70W. Perfect as super tweeter or for autosound use.	T
100			80W; Wedge & Hinged Wedge Available; Introductory Special	\$6.00 each \$22.00 each
100			80W; Wedge & Hinged Wedge Available; Introductory Special	\$26.00 each
22		oy dome tweeter, 110mm metal flange, 72mm cut out, Fs		\$48.00 cach
192	Seas 25TF H456 1" Damped Textile	dome tweeter; 104mm glassfibre/plastic flange, 72mm cu	t out, Fs 1000Hz, 6ohm, 91.5dB, 50W, 26mm VC	\$12.00 cach
5			Fs 49011z, 80hm, 93dB, 100W, metal grill,, fabric surround	\$75.00 each
65			ame, Fs 61Hz, Qis .4, Vas 5.8ltrs, 90dB, 30W, 92mm cut out	\$19.00 each
118			Hz, Qts .37, Vas 6ltrs, 89dB, 50W; 4-1/4" cut out, 2" depth	\$14.00 each
156		E 6.5" paper cone cloth surround woofer, 8 ohm Fs 69Hz	Vas 37ltrs, 100W, 89dB, Very flat response to 4.5K, Good Buy!	\$28.00 each
36		one rubber surround woofer; 80hm, Fs 35, Qts .35 Vas 33		\$7.50 each \$25.00 each
160			n, 6.5mm X-max peak, Fs 27, Qts .3, Vas 75ltrs, 89dB, 150W	\$44.00 each
50			90.5dB, 150W, 40mm Kapton VC, Used in Aria 3 design.	\$85.00 each
232			VC, 6.5mm X-max peak, 180W, F3 of 45Hz in 1cf vented box	\$35.00 each
140			7.74, Qes .34, Qts .33, 92dB, 40oz magnet, 2" VC, 150W	\$25.00 each
33			as 92.4ltrs, 130W, 89dB, Fast and accurate bass response.	\$80.00 each
5 233			/as 46ltrs, 98dB, 100W,8 ohm, high quality bass-midrange.	\$70.00 each
420		ne foam surround; suitable for sealed enclosures, 25W; p-		2 for \$15.00 \$20.00 each
160		cloth surround; Fs 40; Vas 46 ltrs; Qts .33; 91dB; 12Ω; 75 cloth surround; Fs 41; Vas 42 ltrs; Ots. 325; 92dB; 60; 76	Watts; Excellent High Output D'Appolito wooter. Watts; High efficiency, for 2-Way Hi Fi or stage monitors.	\$20.00 each
575		0" Exponential; Fs 1500; 100dB; 16Ω; use a 12dB x-over		\$20.00 each
			at 3K; 3 to 17.5K; Exceptional value. butput systems at reasonable prices, give us a call. This is a rare	
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**PHOTO 3:** Positioning the top piece on the front and back panels.

#### Continued from page 14

of 23" results in a null frequency of almost 300Hz due to the series of 180° bends. A horn with more depth would push the null frequency below 300Hz, which I wanted to avoid. I thought that with the corners properly mitered, the null effects for the configuration could be minimized. The mass rolloff at 411Hz would provide both a graceful attenuation of the response above 400Hz and good mating with a midrange horn above 500Hz.

The Monolith can be assembled from two sheets of 34'' plywood, MDF, or particleboard. I used birch veneer plywood to reduce the weight for shipping. *Figure 10* shows the part shapes, with a plywood cutting guide in *I*-ig. 11. The constant-width box design allows for a certain amount of cutting efficiency. If your lumberyard has a good table or plywood saw, have them make the 34" and 27" cuts for easier transportation back to your shop. You can then make the smaller cuts on your table saw. I also recommend redrawing to full scale the plans in *I*-igs. 8 and 9 as a check on sizes and angles. You can also trace templates for the angles from these drawings.

I first built the horn using only screws, but reassembling it with screws and glue seemed to provide better damping. When I say "attach



PHOTO 7: Duct reflector "D" installed.16 Speaker Builder / 6/93



**PHOTO 4:** Inside view of the throat manifold with attached front and back panels.

piece A to B," therefore, you have the option of either using screws and glue or screws only. Begin by cutting the partition pieces (A, B, C, and J) and assembling them to the top piece, as shown in *Fig. 8* and *Photo 2*. Draw out to full scale the throat manifold on the top piece to ensure your angles are cut correctly. Once the throat partition pieces are attached, you can cut the throat opening with a sabre saw. A router with a long, flush cutting head will do a nice job of trimming the throat port flush to the edges of A, B, and C.

Stand the top piece with the throat partitions on the edges of the front and back pieces (*Photo 3*), using corner clamps to hold them together while you attach the top to the front and back. Then add the two side panels, as shown in *Photos 4–6*. Once you have finished assembling the box, finish joining the throat partitions to the back, front, and side panels. Fill any gaps between the throat partitions and sides with a caulking material such as mortite or silicone rubber.

The duct reflector (D) must be fitted next. This procedure is outlined in the Show Horn article. You first determine the compound angle by cutting and fitting scrap pieces which have the same widths and 45° angles as piece D. Once you have determined the proper saw blade and miter gauge angles, make the same cut on D longer than that specified in Fig. 10. Keep trimming D until it just fits. Attach the duct reflector to the back, top, and side. To further aid construction, you can fit little 45° trianglar pieces under D for alignment and attachment points. Next, redraw the locations of the throat partition pieces on the midpiece. Slide the midpiece into the box and blind screw it to the throat partitions, then fasten the sides, back, and front to it from the outside.

Make some triangular alignment pieces from scrap stock (*Photo 9*). Attach piece I to one of the H reflectors and set it aside (*Fig.* 9). Slide the other H reflector down into the box and attach it with screws to the midpiece, sides, and front.

With more scrap material, cut the triangular pieces that align panels F and G to the correct angles, then attach them on a flat



PHOTO 5: Attachment of the side panels.



PHOTO 6: Completed horn box.

surface. While the structure is still flat, mount the E reflector to panel F, as shown in *Fig. 10*. Using alignment blocks for E is optional. Place the box on its back and slide the divider structure (F, G, and E) into it. Be certain reflector E is touching the midpiece. Attach the divider structure with screws from the sides, then turn the box over and attach it with screws to the back (*Photo 11*). Attach the H-I reflector to the bottom (*Photo 12*). Finally, install the throat reflector.

#### MAKE A STAND

I constructed the stand from square  $1\frac{1}{2}$ " stock, which I cut by ripping up scrap 2 × 4s. Begin the assembly by clamping the top frame on a flat surface, then screw it together with  $3\frac{1}{2}$ " wallboard screws. Repeat the procedure for the bottom frame, and join the frames with the leg pieces. I also added corner braces to correct warping (*Photo 13*). The parts are shown in *Fig. 11*.

Continued on page 18



PHOTO 8: Midpiece installed.



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PHOTO 9: Reflectors (H) with triangular alignment pieces.

G

PHOTO 10: Internal divider structure.

#### Continued from page 16

I determined the stand's height with the best response from a near-field response test, using a Sennheiser MD-421 microphone laid on the floor in front of the horn mouth and a Spectrum Dynamics FFT analyzer operating in an averaging mode with a white-noise input to the speaker. A  $10\frac{1}{2}$ " stand from a previous project gave me a first trial response. I then shimmed up the horn with a series of scrap  $3\frac{4}{4}$ " planks and  $2 \times 4s$  and measured the response at each height. The most balanced response occurred with 3" spacing on the stand, for a total height of  $13\frac{1}{2}$ ".

Figure 12 shows the response with the horn on a  $13\frac{1}{2}$ " stand in a wall position. This "warts and all" response shows much structure in the expanded plot, which is within the  $\pm 3dB$  standard. Above 300Hz, the 3dB drop is probably due to the effects of the 180° bends and the 23" internal depth, as discussed earlier. The 3dB drop is minor, however, compared to the 20dB hole I experienced in the earlier design.

A one-eighth-sized horn is really intended for corner placement. *Figure 13* shows the white-noise response for a corner height of 21". Below 200Hz, there is an increase of approximately 3dB above the wall position response of *Fig. 12.* Above 200Hz, the response is quite ragged compared to the wall position response. The horn apparently needs a mouth reflector to smooth out the reflections. *Figure 13* includes the response as measured by an AudioSource octave-bandpink-noise analyzer to demonstrate how deceptive such a coarse resolution measurement can be.

While the white-noise/FFT-averaged response technique works quite well in the midband, it loses resolution at the low end. For example, Fig. 12 shows that the response at the flare frequency is down some 10dB from the midband bass. Since listening tests did not indicate any serious bass deficiencies, I doubted the validity of the white-noise test at the low end. The Spectrum Dynamics FFT has a transient capture mode, so I decided to use it with a bass pulse. I didn't have a suitable pulse generator, but I found an isolated bass drum pulse on a Telarc CD (#80038, track 3) that made a nice alternative signal. As Figure 14 shows, the bass drum pulse excites the horn's response down to and just below the 40Hz flare frequency. The response's coarse structure is probably caused by comb filtering in the FFT.

As I looked at the Monolith stored on its

side in the corner, I realized that this was the same configuration used by the "Fold and Staple Bass Horn" project, where the mouth exhausts onto the wall.6 I decided to try it, and Fig. 15 shows the response plots for several separation distances from the corner. As you would expect, a 5" separation reduces the output significantly, but the spectrum is markedly more balanced compared to the corner response of Fig. 13. Used in the British "Impulse Horn" design, this narrow mouth loading appears to work from a measurement standpoint ("A New Hope," SB 4/89, p. 65). A quick listening check tended to favor the larger mouth sizes, although more careful evaluations need to be done. For larger separation distances, the primary differences are in the ripple or standing-wave patterns.

#### **BACK CHAMBER**

One of this design's nicest features is the easy-to-access back chamber. I tried a number of test boxes to find two test volumes that would resonate at frequencies below and above the 40Hz flare frequency. To determine the resonant frequencies, I used a signal generator driving the speaker with a 1k $\Omega$  resistor *Continued on page* 24



PHOTO 11: Divider structure in place.18 Speaker Builder / 6/93



PHOTO 12: Final reflector installed.

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## **Industry Standard**

Since its introduction in 1987, *MLSSA* (pronounced "Melissa") has become the loudspeaker industry's standard measuring tool as recognized by the world's leading loudspeaker designers and manufacturers. *MLSSA* is also the system chosen by auto makers, academicians, recording studios and government agencies for many other applications including room acoustics and speech intelligibility measurements.

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## Noise Immunity

Most acoustical measurements are made in environments subject to significant levels of interfering background noise. Nonstationary noise, such as impulse noise due to doors closing or, sporadic noise due to local traffic can result in severe errors with some other measuring systems. A significant advantage of MLSSA is that all nonstationary interfering noise, whatever its source, is automatically converted to stationary noise during the measurement process. Stationary noise is much more benign than nonstationary noise because it is spread out evenly over time and is largely windowed away in loudspeaker frequency response measurements. In some room acoustics measurements, MLSSA provides postprocessing algorithms capable of removing even the residual stationary noise.



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## **Fast Measurements**

Although *MLSSA* contains no DSP chip, it is actually faster than some more costly DSP-based systems when running on the faster personal computers. Eschewing an integrated DSP chip not only lowers hardware costs but also forestalls obsolescence by allowing *MLSSA*'s performance to improve in step with improvements in PCs. Running on a 33 MHz 486, for example, anechoic loudspeaker frequency response measurements including acquisition, computation and display require only 3.1 seconds using a full-length 64K-point MLS stimulus or, just 1.6 seconds using a shorter 32K-point sequence. Altogether, four distinct sequence lengths of 4K, 16K, 32K and 64K-points are provided in hardware for high precision, low jitter MLS generation.

## **Enormous Time-Bandwidth Product**

A third fundamental advantage of *MLSSA* is its enormous time-bandwidth product. This feature is especially useful in room acoustics and speech intelligibility applications because it allows you to measure a long impulse response over a wide bandwidth quickly and store the results to disk. Then, through its powerful post-processing functions, software can perform detailed narrowband analysis later from stored time-domain measurements. But this feature is equally useful in impedance measurements of drivers or loudspeaker systems. A time-bandwidth product of 20,000 allows you, for example, to measure impedance over a 1 kHz bandwidth with 0.06 Hz resolution or, over a 20 kHz bandwidth with 1 Hz resolution.

## **Room Response**

*MLSSA* measures and displays room impulse response curves up to 65535 points in length with a maximum time-bandwidth product of 20,000. This allows a 4 second impulse response over a 5 kHz bandwidth or, a 2 second response over a 10 kHz bandwidth. Room impulse response measurements can be stored on disk and later retrieved for further octave-band or other analysis. Room impulse response measurements can also be exported.



Overlay Calculate Printer DOS Units Library Info Quit P1 for Help MLSSA: Time Domain

By applying a large FFT to the room impulse response, MLSSA determines the room frequency response which can then be smoothed to 1/1 or 1/3 octave (see figure opposite). FFT sizes ranging from 32 to 65536 points are supported.

## Math Operations and Spatial Averaging

*MLSSA* can perform many mathematical operations on both time and frequency data files including average, add, subtract, multiply, divide, smoothing, convolution, correlation and the inverse FFT. Many mathematical operations also accept both real and complex constants for scaling operations or other purposes.

Both complex and power averaging are provided for frequency data files. Power averaging is required for correct spatial averaging of room frequency response data. Room impulse response measurements can also be averaged to determine spatially-averaged acoustical parameters such as reverberation time.





## Fast Equalizer Adjustments

Currently used for final test and equalization of high-end home theater installations, *MLSSA* simplifies and speeds equalizer adjustments in equalization applications. After measuring the smoothed and spatially-averaged room response, simply connect *MLSSA* to measure your graphic or parametric equalizer in nearly real time. Then you merely adjust the equalizer for the desired target frequency response. Adjust equalizers quickly, as a simple post-processing operation which is much faster than trying to repeatedly remeasure the average room response.

## Schroeder Plots and Reverberation Time

A Schroeder plot is the reverse integral of the squared room impulse response and reveals reverberant decay of the sound field. When graphed on a logarithmic scale, its slope is a measure of reverberation time. As a post-processing operation, MLSSA can usually remove the influence of both stationary and nonstationary background noise from Schroeder plots for improved accuracy. MLSSA not only calculates wideband Schroeder plots and reverberation times but can derive them for any desired frequency band through its built-in programmable digital bandpass filter. This bandpass filter permits reverberation measurements to be made over any octave or fractional octave band simply by post-processing the measured wideband impulse response. Reverberation time is determined from Schroeder plots using linear regression over selected decay ranges.

## **Energy-Time Curves**

*MLSSA* computes both wideband and narrowband (filtered) ETCs as the envelope of the measured impulse response with a choice of frequency domain window functions. Computes unwindowed and half-Hann ETCs as recommended by John Vanderkooy and Stanley Lipshitz of the Audio Research Group. ETCs can locate room reflections or, assist in time delay adjustments of loudspeaker clusters. The figure below shows a wideband noise-corrected Schroeder plot (dotted curve) displayed simultaneously with the wideband energy-time-curve (solid curve).



TIME DOMAIN MEMU: Go View FFT Waterfall Acquisition Setup Transfer Macro Overlay Calculate Printer DOS Units Library Info Quit F1 for Help MLSSA: Time Domain

## **Tabular Acoustical Parameters**

*MLSSA* computes reverberation time (RT) as well as other important acoustical parameters automatically in IEC-standard octave bands (see figure opposite). Computes C50, C80, EDT and Ts as well as RT for four selectable decay ranges and displays the results in tabular form for easy interpretation and documentation. Room noise correction is automatically applied to EDT and RT measurements for high accuracy.

## **DAT Machine Support**

A special Asynchronous Cross-correlation acquisition mode makes possible remote room measurements using any DAT machine. DAT machines can used to pre-record the MLS stimulus or, to record the raw room response data for later analysis by *MLSSA*. This eliminates the need for long cables when making measurements in large spaces.

## Absolute SPL and NC Ratings

*MLSSA* measures sound or noise levels in dB-SPL in IEC-standard 1/1, 1/2, 1/3, 1/6 octave bands. Determines the NC (noise criteria) rating of noise sources and calculates A, B, and C weighted dB-SPL.

Band	3	4	5	6	7	8	9
Parameter	125	258	580	1000	2008	4808	8868
S [dB-SPL]	71.7	75.6	78.6	81.0	82.6	84.3	83.1
N [dB-SPL]	54.5	56.0	55.8	59.2	60.1	61.0	58.4
S/N [dB]	17.2	19.6	22.8	21.9	22.6	23.3	24.7
C50 [dB]	-8.48	-5.57	5.87	-5.88	-3.32	-1.33	1.35
CB0 [dB]	-7.17	-4.43	-3.32	-3.52	-1.52	0.30	2.96
[an] at	237.8	232.4	215.2	219.1	184.5	139.8	99.7
[2] TCE	2.952	3.420	3.279	3.185	3.169	2.647	1.916
RT-1 [s]	2.489	2.682	2.889	2.659	2.653	2.477	1.826
(-5,-20) r	-0.989	-0.997	-0.991	-0.997	-0.998	-1.888	-8.999
RT-2 [s]	2.300	2.426	2.628	2.411	2.451	2.369	1.855
(-5,-25) r	-0.990	-0.992	-0.983	-0.989	-0.993	-0.998	-0.999
[2] E-TR	1.728	1.985	1.967	1.989	1.939	2.067	1.869
(-10,-30) r	-0.980	-0.961	-0.971	-0.976	-0.976	-0.989	-1.888
RT-4 [s]	1.234	1.158	1.245	1.321	1.326	1.616	1.778
(-15,-35) r	-0.937	-0.933	-0.941	-0.959	-0.955	-0.972	-0.999

IDC Octave Band Acoustical Parameters

ESC to exit or F1 to print.

MLSSA: Acoustics

## **Energy-Time-Frequency Plots**

Reverberant decay can also be visualized through the 3D energy-time-frequency plot (see figure below). *MLSSA* also provides this function complete with 3D cursor readout.





ESC to exit, F1 to print, F2 and cursor keys move cursor MLSSA: Waterfall

## Speech Transmission Index

Based on the modulation transfer function (MTF), the speech transmission index (STI) is an objective measure of speech intelligibility, widely recognized as being highly accurate in predicting subjective speech intelligibility. MLSSA computes full STI from the measured impulse response and, properly accounts for all contemporaneous interfering background noise whether stationary or nonstationary noise. MLSSA has recently been approved for use by aircraft companies to measure the STI of cockpit voice flight recorders.

**HTF Hatrix (Calibrated)** 

Frequency-Hz	125	258	588	1888	2000	4800	9999
level 4B	-24.5	-22.9	-23.2	-24.8	-26.4	-32.2	-65.5
a-correction	1.000	1.000	1.000	1.000	1.000	8.999	0.685
0.63	0.988	8.987	8.977	0.936	0.791	0.552	0.071
8.80	8.982	8.984	0.975	8.933	0.794	8.542	0.067
1.00	0.982	8.984	8.975	0.933	0.794	8.542	0.067
1.25	0.978	8.982	0.974	0.935	8.793	0.553	0.068
1.68	0.978	0.982	0.974	0.935	0.793	0.553	0.068
2.00	0.973	8.979	0.973	0.936	0.798	0.548	0.068
2.50	8.967	0.976	0.972	0.933	0.793	0.545	0.052
3.15	8.958	0.971	0.968	0.930	0.791	0.558	6.658
4.88	8.946	8.967	0.966	0.929	0.787	0.535	8.647
5.00	0.935	0.961	0.962	0.929	0.793	0.536	0.060
6.38	0.920	0.952	0.958	0.925	0.787	8.544	0.072
8.88	0.965	0.946	0.953	0.923	0.786	8.555	0.074
10.00	8.888	0.933	8.947	0.920	0.796	8.549	0.059
12.50	0.854	0.920	0.948	8.918	0.786	0.551	0.057
octave TI	0.926	0.967	0.975	0.874	0.693	0.528	0.109



MLSSA's STI function properly accounts for both interfering background noise and nonlinear distortion based on a single measurement, provided only that the MLS stimulus is first passed through a simple speechweighting filter prior to applying it to the system to be analyzed. This important feature was previously available only on costly dedicated RASTI instruments.

In addition to rooms, MLSSA also correctly measures the STI of digital devices, such as Codecs, as well as analog tape recorders which include a tape monitor function.

For STI measurements performed with a human mouth simulator, software correction of errors in the simulator's frequency response is also supported. MLSSA also estimates the percentage articulation loss of consonants or %ALcons from the measured STI value. Running on a 33MHz 486 computer, a full STI analysis is completed in a mere 4 seconds.

## Rapid Speech Transmission Index

MLSSA can also measure the less accurate but internationally standardized rapid speech transmission index (RASTI) according to IEC 268-16. Like STI, the RASTI function also properly accounts for interfering background noise provided a speech-weighting filter is present.

125	258	500	1866	2868	4800	8888
		-23.1		-26.2		
				0.786		
		8.979				
				0.797		
		0.980				
				0.890		
		0.978				
				0.797		
		0.976				
				0.797		
		1				
	125	125 258		-23.1 0.979 0.989 0.978	-23.1         -26.2           0.979         0.796           0.980         0.896           0.978         0.797	-23.1 -26.2 0.979 0.989 0.997 0.989 0.896 0.896 0.896 0.897 0.896 0.897 0.896 0.797

HTF	Natrix	(Cal	ibrated
	1100 21 17		

RASTI value= 8.831 ALcons= 1.9% Rating= EXCELLENT

ESC to exit or F1 to print.

MLSSA: RASTI

## Modulation Transfer Functions

MLSSA's also makes wideband or fractional octave MTF measurements for more detailed analysis than is possible with STI alone.



PREQUENCY DOMAIN MEMU: Go View Reference Acquisition Setup Transfer Macr Overlay Calculate Printer DOS Units Library Info Exit **F1** for Help MLSSA: Prequency Domain

## **Microphone Calibration and Correction**

You can enter and store microphone sensitivity data on up to 10 microphone/preamp combinations. For higher accuracy, *MLSSA* will optionally calibrate your microphones using an external microphone calibrator. *MLSSA* will also correct out minor deviations from flat response by importing microphone frequency response data taken from its calibration curve.

## Autorange and Programmable Bandwidth

*MLSSA* automatically adjusts its input gain on each measurement for maximum dynamic range. Input levels from 10 millivolts to 20 volts RMS are easily accommodated. Autorange can also be disabled and the input gain set manually if desired. Measurement bandwidth can also be programmed anywhere from 1 kHz to 40 kHz through *MLSSA*'s high-quality 8th-order on-board antialiasing filter.

## Logarithmic or Linear Data Export and Import

*MLSSA* will export data in a standard text file format for use by loudspeaker CAD packages such as *CALSOD*, *LEAP* or *XOPT*. Exports frequency domain data in either linear format or, in logarithmic format for reduced file size. In the frequency domain, you can import both text and binary data files. The imported data is automatically interpolated by cubic splines to match the frequency spacing of the currently displayed curve allowing you to overlay curves having different frequency spacings.

## Integrated Macro Processor

*MLSSA* contains an integrated macro processor. A complex series of *MLSSA* commands can be recorded as a macro and then played back later through the action of one or two keystrokes. No programming skills are required to create macros. Special macro commands permit remote initiation of measurement cycles or, they can be used to synchronize *MLSSA* with a motorized turntable for automated polar response measurements.

## **Autonamed Filenames**

*MLSSA* provides a automatic filename system to keep your measurement data organized. Whenever you save data to a file, *MLSSA* will optionally create a new numbered filename and save that file to any preselected drive and directory. *MLSSA* also maintains an audit trail for all measurements saved to disk files because all measurement setup parameters are automatically stored along with the measurement data.

## **Built-in Screen Capture**

A integrated utility captures any graphics screen in full color to a PCX file. Used for importing graphics into desk-top publishing (DTP) packages or for future redisplay by *MLSSA*. Text screens are captured to plain text files for exporting STI, RASTI or the tabular acoustics screen.

## Printer Support

*MLSSA* offers direct support for HP LaserJet, IBM Graphics, Epson and Okidata printers. You can also redirect printer output to a file when a printer is not available. Later, *MLSSA* can print the contents of the file. An optional header or footer file can also be attached to all graphics printouts, including waterfall plots. These files can contain anything including company logos or product information.

## **Computer Requirements**

MLSSA consists of a full-length card and software designed to run on standard personal computers running the MS-DOS operating system versions 2.1 and above. The card is compatible with all XT (8-bit), AT (16-bit) and EISA (32-bit) full-length expansion slots.

Your computer must contain a math coprocessor chip except on 486DX machines which already include the math coprocessor on the main CPU chip. Your computer must also include CGA, EGA or VGA graphics, a hard disk and, at least 640 kilobytes of memory. For optimum performance a 386SX computer or better is recommended having at least 2 megabytes of memory, VGA color graphics, MS-DOS version 5.0 or better and, an HP LaserJet series III, 4 or compatible printer.

## **Ordering Information**

A complete *MLSSA* system includes a 1 year hardware warranty and free software updates for the first year after purchase. For a demo disk, current prices or, a list of authorized overseas distributors contact:

> DRA Laboratories 24 Halifax CT Sterling, VA 20165 USA Tel (703) 430-2761 Fax (703) 430-0765



PHOTO 13: Finished 131/2" stand.

#### Continued from page 18

in series and an AC voltmeter across the speaker input. I then plotted the resonant frequencies (Y axis) and their corresponding volumes (X axis) on linear graph paper and drew a straight line between the two points. At the point where the line crossed the 40Hz frequency, I read the required volume off the X axis. With a table saw, I trimmed down the larger of the two test volumes to the required volume. Rechecking the horn/back volume system's resonant frequency showed it very close to 40Hz.

I found the resultant back chamber volume to be 1.81 ft.<sup>3</sup>, corresponding to an enclosure with external dimensions of  $17V_2'' \times 17V_2'' \times$ 13". If I had it to do over, I would make the square dimensions slightly rectangular. Feel free to experiment. I calculated the theoretical back volume to be 1.17 ft.<sup>3</sup> (based on annulling the throat reactance of an infinite horn by a back volume). Assuming a driver volume of 0.2 ft.<sup>3</sup>, the experimentally determined back volume is 38% larger.

In my experience, the experimental back

volumes for one-eighth-sized horns are always larger than the theoretical ones, sometimes by as much as a factor of two. The fact that the Monolith's back chamber is off by only 38% indicates that it is close to the ideal infinite horn. For one-quarter-sized bass horns, the back chamber is usually much closer to the ideal limit—but at the expense of a much larger horn enclosure.

I used  $V_2$ -inch-wide foam rubber weather stripping to seal the edge of the back chamber where it mates to the top of the horn. To confirm that reactance annulling actually works, connect a signal generator set at 40Hz directly to the speaker input with the back chamber in place. Adjust the generator output until you can hear the 40Hz signal. Since the horn is very efficient, you don't need an amp for this experiment. Pop the back chamber seal by slightly prying up one side of the box with a screwdriver. As soon as the air seal is popped, the 40Hz signal cannot be heard; restore the air seal and the signal returns.

The back chamber can be firmly attached to the top of the Monolith with L-brackets, or

with the more decorative copper-colored chair brackets. I also attached a speaker mounting plate to the top for added mass and strength. You can compensate for the decrease in volume with acoustical stuffing in the back volume.

### INTEGRATION AND PERFORMANCE

When I started this project in 1988, the 2220 was a staple of JBL's pro line, but they discontinued the model in late 1992. Some surplus units may be found on the used market. but other models will also work. The EVM15L is the closest to the 2220 with a mass rolloff above 400Hz; the "B" version will have a lower mass rolloff near 300Hz. Drag out any of your old 15" drivers and try them on the Monolith horn as an interesting experiment. Those designed for direct radiator applications will sometimes work well, but experiments show that a horn needs a driver with a big magnet and light cone. In addition, many different drivers will resonate near 40Hz with the same back chamber.

I compared the relative pink-noise responses between a well-calibrated midrange horn and the Monolith with the 2220 driver. The Monolith in a wall position appears to have a sensitivity of about 105dB, measured with the horn positioned against a wall but close to a corner (4'). Against a long wall, the apparent sensitivity can be lower.

The Monolith appears to be less directional than the Show Horn, because the mouth exhausts onto the floor. This broad radiation pattern can impact the Monolith's integration with a midrange horn. After some experimen-

### SOURCE

Bright Star Audio 2363 Teller Rd. #115 Newbury Park, CA 91320 (805) 375-2629 FAX (805) 375-2630



FIGURE 12: Monolith 40Hz horn response with a  $13 \nu_2''$  stand, wall position.



FIGURE 13: Monolith 40Hz horn response with a 21" stand, corner position.





FIGURE 15: Response of the Monolith horn in an upright position for several separation distances from the corner.

tation, I found that the Monolith bass and midrange horns integrated best when the latter's radiation pattern is the broadest in the horizontal plane. For most rectangular-mouth midrange horns, this occurs when the horn's longest axis is horizontal. For the Show Horn, which has a narrower radiation pattern in the horizontal plane, the midrange horn's long axis must be arrayed vertically so the radiation patterns match.

The Monolith stands high, so I do not recommend mounting the midrange horn atop the bass horn—it needs to be at ear level. I usually mount the midrange horn in a separate box and place both it and the tweeter on a 2' spiked stand in front of the bass horn, as shown in *Fig. 16*.

Finally, you must tune the Monolith. This may be the first time you have heard anyone refer to "tuning a horn" other than adjusting the back chamber system resonance. My friend Barry Kohan of Bright Star Audio has taught me about damping speaker systems with sand loading. The improvements can be dramatic, especially in the bass. One of the problems with the throat manifold design is an apparent multiple backwave reflection region around the back duct opening (*Fig. 8*). Play any CD with a tremendous bass pulse and place your hand on the top of the Monolith over the back duct. You will definitely feel the wood give and rise up in response to the pulse.

You can damp out this enclosure flexing with mass damping, using a Bright Star "Little Rock" isolation pod on top of the Monolith next to the back chamber (*Fig. 17*). An alternative is to buy several sacks of lead shot and place them on top of the throat manifold. The effect can be dramatic. You can also buy a Bright Star "Big Foot II" sand-filled isolation base and cover the rest of the top plate. A second improvement can be made by loading the top of the back chamber with sand or lead. A "Little Rock" pod or a single bag of sand or lead shot will work.

**ACKNOWLEDGEMENTS** 



FIGURE 16: Monolith horn with midrange horn configuration.

I thank Fred Ireson for suggesting the project. His enthusiasm for hom loudspeakers kept me on track. I also thank Manfred Buechler for his excellent photos.

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 Davis, N., "Fold and Staple Bass Horn," Audio Amateur Loudspeaker Projects, Marshall-Jones, 1985, p. 103; The Audio Amateur (Issue 1, 1977): 10–13, 20. People have asked me about the advantages of a horn design going down to 40Hz. For many types of music, it doesn't make much difference, but the 40Hz horn does resolve the low end better than a 50Hz horn.

The Monolith horn was a learning project with two main lessons: understanding horn folding problems and sand loading. The effects of 180° bends can be neutralized somewhat by intelligent placement of diagonal reflectors. Sand loading can damp out multiple reflection problems and improve the bass quality. The 40Hz Monolith horn project is complicated, but any good wood craftsman should find it well within his means. Its modular design and construction should lend itself to much experimentation.



# ORBITING SATELLITES

### **By Bill Fitzmaurice**

Changes in living space forced me to change my speaker setup. Specifically, the acquisition of a large entertainment center left me with no home for my 15-year-old corner horns. Between new furniture and some other remodeling, I just had no place for my faithful Klipsch clones.

A subwoofer-satellite setup I had used some 25 years ago seemed the reasonable way to go again, but even conventional satellites mounted on the wall at listening level wouldn't fit my available space. I finally decided on a mounting scheme that would virtually allow me to launch my satellites into orbit: at the junction of the wall and ceiling, as high as possible without blasting through the roof (*Photo 1*).

This design offers a few advantages over standard practice. The satellites are removed from kids and pets, so protective grilles are unnecessary. The downward-firing angle ex-



FIGURE 1: Parts layout: a. top view; b. side view; c. baffle front view.

pands the depth of the listening area, and the triangular shape eliminates standing waves and conserves on expensive materials. As a bonus, the cones don't even collect dust.

I resisted the temptation to spend more money than I had to achieve the desired result. For drivers, I chose Pioneer  $4\frac{1}{2}''$  woofers and a generic titanium-dome tweeter.

The small amount of wood required made solid wood economically feasible and vastly simpler than veneer. I used solid oak for the exposed surfaces and  $V_2''$  baltic birch plywood for the remainder. The unexposed pieces can be plywood, MDF, or whatever you have on hand.

#### ASSEMBLY LINE

Simultaneously cut the top, top liner, back, and baffle to width on a table saw, so the rip fence setting will remain unchanged and all the parts will be identical. When cutting the top to length, using anything other than  $\frac{1}{2}''$  plywood will cause an alteration in the part's length. For instance,  $\frac{3}{4}''$  material will result in an overall top length of  $111\frac{4}''$ .

Cut the top for the crossover mounting (*Fig. 1*). Glue and screw the liner to the inside of the top, then attach the top/liner assembly to the back.

Cut the driver holes in the baffle and drill it for the driver mounting T-nuts. Prefit the



PHOTO 1: Completed cabinet, front view.





baffle in place, checking that the woofer magnets don't hit the top or back. If they do, rout out sufficient clearance as necessary. Now you can glue and screw the baffle to the cabinet. You'll have a slight overhang at each end which is trimmed flush, as in *Fig. 2*.

Using the top/back/baffle assembly as a template, trace out the sides on the  $1 \times 10$ s, leaving at least 1/8" of selvage all around. Continued on page 28

TABLE 1						
S	SATELLITE PARTS LIST					
Qty. Crossover	Description					
2 1 1	0.9mH coils 0.3mH coil 0.2mH coil					
1 1 3	10Ω 10W resistor 47mF capacitor 22mF capacitor 4.7mF capacitor					
Lumber 17" 5' 1⁄2"	$1 \times 6$ $1 \times 10$ Plywood or other material as desired (for top, top liner, back) 5" $\times$ 32"					
Tweeter 1	Titanium Dome (MCM #53-325)					
Woofers 2	Pioneer 41/2" (MCM #51-075)					
Miscellaneous						

Glue, wood screws, T-nuts and bolts for driver mounting, silicone sealant, speaker wire, binding posts or jacks, water-based urethane.

#### Note

All materials are per cabinet. Two cabinets will require doubling materials count.

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PHOTO 2: The crossover mounted in its recess on the cabinet top.

#### Continued from page 26

Simultaneously glue both sides to the assembly. While the glue sets, you'll need three clamps with a minimum of 7" capacity to hold the assembly together: one at each corner. Use a flush-trimming router bit to trim the sides with the top/back/baffle, then chamfer all the edges using a  $\frac{1}{2}$ " radius on the baffle face and a  $\frac{1}{4}$ " radius on the remaining edges.



**PHOTO 3:** Hanger bracket detail, showing recessed bracket mounting.

Drill three 1/4" holes in the top liner for driver wiring, and whatever holes are required by your choice of binding post/jack. If you plan to inlay your wall hanger on the back, rout it out as well.

To finish the cabinets, I used a water-based urethane. After first sanding them smooth with 100- and 220-grit papers, apply one coat. Within two hours, sand again with 220-grit paper and apply the finish coat. Water-based urethanes do not impart any color to the wood, so you must stain to match your furniture.

Wire up your drivers and mount them to the baffle. Run separate wires from the woofers and the tweeter through the top liner. Be certain to label them and note their polarities. Seal the drivers to the baffle with silicone sealant; seal the wiring holes, as well. Depending upon your jack/binding post arrangement, you may also need to run wires from the posts to the inside of the cabinet and then back out again through another hole. Do this



FIGURE 3: Crossover schematic/wiring diagram.

before sealing the drivers in place. With all the wiring on the exterior, you can wire up your crossover and test or modify it without reopening the cabinet.

#### **BOTTOM LINE**

I began my design with a simple 18dB crossover at 3.2kHz, and wired my woofers in series rather than parallel (*Fig. 3, Photo 2*). The resulting  $16\Omega$  impedance lowers woofer response relative to the tweeter, which compensates for the woofers' corner loading.

As you might expect, the original response curve is fairly ragged with a peak in the crossover area. To smooth it, I added a bandpass rolloff filter to the tweeter. This cut the undesirable peak from 2.4–6.4kHz without *Continued on page 71* 



8

PHOTO 4: A satellite "in orbit" above the enter -

FIGURE 4: Frequency response, measured with Radio Shack sound-level meter at  $V_2m$ , on tweeter axis. Plot corrected for meter high-frequency rolloff.

tainment center.

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## REAL-WORLD Three-Way Crossovers

By Ralph Gonzalez

In my article "Real-World Two-Way Crossovers: A Design Method," (SB 2/92 p. 18) I described "variable-Q" formulas for varying the amplitude at the corner frequency of second-order high- and low-pass crossovers. Varying this parameter helps account for the bandwidth limitations and phase shifts present in real-world drivers. I mentioned in the article that more complicated formulas exist for second-order bandpass (midrange)

crossovers. Figure 1 is a BASIC computer program which uses these formulas.

You enter the low and high crossover frequencies and their respective Q values, as well as the load resistance (including the effects of attenuating resistors). The program then prints the capacitor and inductor values for a second-order bandpass crossover (*Fig.* 2) with the desired characteristics. An example of the filter response with  $f_L = 500$ ,  $f_{H} =$  4,000,  $Q_L = 1$ ,  $Q_H = 0.350$  is shown in *Fig. 3*. This sort of response may be useful to equalize a midrange driver which has an overdamped low end and a peaky high end.

Recall from my previous article that an ideal two-way Linkwitz-Riley second-order crossover has Q = 0.5 for both the high- and low-pass sections. (This means that the woofer and tweeter amplitude is -6dB at the crossover frequency, producing a flat overall

#### PI = 3.1415926

```
INPUT "Enter LF and HF corners (Hz)";fL,fH
S = fH/fL
IF S <= 2. THEN
    PRINT "Too narrow spread"
    ST0P
END IF
a = 2.*(S-1.)/SQR(S^2-2.*S)
PRINT "(Q for ideal APC crossover:";1./a;")"
INPUT "LF and HF Q";QL,QH
H = S + a^2 - 4 + 3./S
B = S + a^{2} + 1./S
K = B - 1.
EG = 20.*LOG(K/H)/LOG(10.)
PRINT "(Excess gain of ideal APC crossover:";EG;")"
INPUT "Enter total load resistance (Ohm)";k0
wL = 2.*PI*fL
wH = 2.*PI*fH
xL = 1./2./QL
```

```
xH = 1./2./QH
```

FIGURE 1: Program listing. Finds components for second-order bandpass crossover, given the user's choice of corner frequencies and their





#### b4 = 1./wL<sup>2</sup>/wH<sup>2</sup> b3 = (2.\*xL\*wL + 2.\*xH\*wH)/wL<sup>2</sup>/wH<sup>2</sup> b2 = (wL<sup>2</sup>+wH<sup>2</sup>+4.\*xL\*xH\*wL\*wH)/wL<sup>2</sup>/wH<sup>2</sup> b1 = (2.\*xL\*wL\*wH<sup>2</sup> + 2.\*xH\*wH\*wL<sup>2</sup>)/wL<sup>2</sup>/wH<sup>2</sup> b0 = 1. 'assume b0=1 in transfer function k1 = b4/b3/k0 k3 = b1\*k0 - (b3<sup>2</sup>\*k0/(b2\*b3 - b4\*b1)) k4 = b3\*k0/k3/(b1\*k0-k3) k2 = b1\*k0 - k3 PRINT PRINT "C1 =";k1\*1e6;"uF" PRINT "L1 =";k2\*1e3;"mH" PRINT "L2 =";k3\*1e3;"mH" PRINT "C2 =";k4\*1e6;"uF"

END

respective Q values. Suggests values for ideal second-order APC as a starting point, based on Robert Bullock's *SB* 2/85 article.



frequency response.) I recommended this value as a starting point for attempting to develop a crossover using real-world (non-ideal) drivers. However, the choice of Q is not so simple in the case of an ideal three-way crossover: it varies according to the spread of the crossover frequencies.

Part II of Robert Bullock's "Passive Crossover Networks" (*SB* 2/85, p. 26) presents formulas for computing, among others, ideal second-order three-way networks. Bullock's "a" parameter is equal to 1/Q in the secondorder case. My bandpass program computes this parameter using Bullock's formula, and suggests the corresponding Q as a possible starting point for both  $Q_L$  and  $Q_{H}$ .

It also shows the excess gain of Bullock's ideal bandpass crossover, the starting point for calculating the values of attenuating resistors (Bullock's RA). If you use the same "ideal" Q in my original formulas for the woofer and tweeter crossovers, and if you reverse the midrange polarity, you will obtain Bullock's three-way all-pass crossover (APC). This crossover will have a flat frequency response-until you use it with real-world drivers.

If measurements or computer modeling indicate that a peak or valley occurs at one or both crossover frequencies in the real-world system, you can use the same approaches l previously suggested to customize the crossover: reverse one driver's polarity, vary the Q of the respective crossovers, or use noncoincident crossover frequencies (such as a 400Hz low-pass filter on the woofer and a 600Hz LF corner on the bandpass filter).

In the case of the bandpass crossover, you may vary  $Q_L$  and  $Q_H$  independently. You may also find it necessary to use a different gain value from that in the "ideal" crossover, by changing the series resistor value. This necessitates recomputing the network values with the new load resistance, unless you use a series-parallel combination to maintain the same overall resistance.

## SOURCE CODE INFO

Copies of this BASIC IBM program are available on  $3V_2''$ 720K DS/DD disk (specify SOF-GON1B3) or  $5V_4''$  360K DS/DD disk (specify SOF-GON1B5) for \$6.95 plus \$3 s/h from Old Colony Sound Lab, PO Box 243, Peterborough, NH 03458-0243; (603) 924-6371, FAX (603) 924-9467.







Reader Service #32

# THE SIMPLEX

## By Ernie Pfannenschmidt

**B**owed stringed instruments produce some of the most pleasing musical sounds imaginable, yet they are just simple wooden boxes brought to resonance by catgut stroked with strands of horsehair—not exactly high technology by anyone's standards.

Since no one has so far come up with a perfect transducer, our loudspeakers may also be considered musical instruments. We can build them, like fine violins, without recourse to technological wizardry. The Simplex (*Photo 1*) is an example of what you can do with inexpensive components in an ordinary enclosure. Coupled with a suitable subwoofer, it rivals systems in the four-figure price bracket but costs under \$200 to build.

The only tools required are common household woodworking tools and a soldering iron. The definitive testing instruments are your own ears and sound perception. You alone must be happy with them; test lab results, favorable or otherwise, are irrelevant.

#### THE DRIVERS

The drivers in this vented, two-way speaker are fed by a biwired passive first-order Butterworth crossover. The woofer is zobeled. I started the project by purchasing two Radio Shack model 40-1296 crossovers that offer a choice of crossover frequency at 2, 2.4, or 4kHz.

l then looked for inexpensive drivers to match the 6dB slope of these crossovers and chose the Radio Shack model 40-1011-A 61/2'' polypropylene woofer. This woofer is well made and moderately priced. Its parameters are:  $Q_{7S} = 0.4-0.46$ ,  $f_S = 38-50$ Hz. The frequency response extends to 5--6kHz and then drops off at a suitable slope. The exact values are not terribly important for this project. Good stereo reproduction does not necessarily require precisely matched pairs of drivers in the right and left enclosures. A change in the Radio Shack catalog number suffixes after 1 bought my woofers indicates a switch in manufacturers and some deviation

### **ABOUT THE AUTHOR**

Emie Pfannenschmidt is a retired engineer who spent much of his career designing scientific instruments with the National Research Council of Canada. He has been an audio enthusiast since the 1950s and enjoys sailing among the islands and along the fjords of British Columbia.



PHOTO 1: The Simplex speaker.

in specifications. The quoted spread is typical and quite acceptable.

*Figure 1* shows a typical impedance sweep of a 40-1011 woofer. I ran some tests to decide what 1 should do about the zobel. 1 soldered the components in series (*Fig. 2*) directly across the driver tabs. If you like to experiment, retain the resistor and try different capacitors to suit your ears.

The 100 DT 26/72 Peerless tweeters are a bargain. They offer sufficient overlap for a possible low crossover point of 2kHz, and their efficiency level matches the woofers'. Most important for a two-way system with a 6dB crossover is that the tweeters have an excellent high-power-handling capacity.

#### THE ENCLOSURE

Few speaker builders will dispute the importance of volume in designing enclosures for the lower-octave regions, yet relatively few seem to give it much thought when addressing the midrange. Miniaturization seems to be in vogue. I prefer a healthy ratio between the driver cone area and the box volume for the midrange as well. The Simplex enclosure (*Figs. 3* and *4*) has an internal net space of 0.82 ft.<sup>2</sup> and a box resonance of about 62Hz. Without consulting Thiele/Small, I guessed at a 11/2-inch-diameter vent duct that was 1-11/2'' long, which worked out just right in the end.

When it comes to soundstage imagery, there is little to fault in vertically and symmetrically aligned driver arrangements. The speaker's front baffle is as narrow as convenient and both beveled and round-edged to *Continued on page 34* 



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FIGURE 2: Biwired crossover for the Simplex speaker.

#### Continued from page 32

reduce diffraction effects and enhance dispersion. The proportions of the box sides relative to one another are based on averages derived from an analysis of especially good sounding vented speakers of similar volume.

Over the years, I have built speaker enclosures of various materials, but I still prefer high-grade plywood, in this case 3/4" 6- or 7-ply birch. It seems to give a natural and less synthesized timbre to the music, particularly when playing some of the "harsher" compact discs. The front baffle is a laminate made of a 3/4-inch-thick air- or kiln-dried spruce board and a 3/8" fir plywood backing. The enclosure is important to the success of this project, so follow the drawings carefully.

To decouple the woofers, rest them on a silicone seal or a custom-cut soft gasket underlay. They should be flush with the baffle face. You can screw the tweeters directly onto the baffle without recessing or decoupling.

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In questioning authors, please leave room in your letter for replies which should relate to the article, be framed clearly, and written legibly. Please do not ask for design advice or for equipment evaluations. Use Elmer's white glue and finishing nails to assemble the enclosure. Pipe clamps hold things plumb and square until the glue cures. The side and back panels have internal braces at different heights. Glue rubber-backed nylon carpeting to the side and rear panels only (*Fig. 3*). The bottom panel is held in place by countersunk flat-head wood screws and is removable, providing access to the inside. Work carefully here and at the vent hole so you get a snug fit. Fill each cabinet evenly and loosely with  $5\frac{1}{2}$  oz. of polyester Fiberfil.<sup>®</sup>

Modify the crossover as in *Fig. 4*. On final assembly, place neoprene washers under the crossover board and fasten it to the enclosure bottom with round-head brass wood screws.

#### TUNING YOUR SPEAKERS

When you're finished building the enclosures, install the drivers, to which long heads have been soldered. Do not install the crossover yet. Have some cardboard vent tubes I-3'' long at hand. Insert either the shortest or the longest of these tubes and feed the speaker wires through to the outside, where you can clip them to the crossover. Also feed the crossover the signal from the amplifier's outputs. Then proceed to "tune" your creations: this is the fun part and will keep you busy many an evening. Use your own amplifier/receiver. Play only discs with which you are familiar, over and over again. They must span the musical spectrum, but remember that these speakers will produce only an acceptable upper bass and no lower bass.

First decide which crossover frequency sounds best under all circumstances. It should be the 4kHz crossover, which lets the woofer supply the necessary power and impact. Placing the crossover dip there moderates tape hiss because this is where it peaks.

Now determine which vent length provides the cleanest bass (not necessarily the lowest). I settled for 11/2" vents, but you must make your own decision based on what sounds best to you. Line the inside of your vent with Neotex, an open-mesh, nonskid, neoprene-coated matting sold by yacht chandlers or at outlets catering to people with physical handicaps. Provided you don't start manufacturing such a vent, you won't violate the Russian patents addressing this way of modifying the backwave. The material also works well as a mat atop speaker stands.

Finally, you may try changing the amount of Fiberfil stuffing in your speakers. Like a luthier tuning a fine violin, you can get it just right with patience.

Continued on page 36





## A better speaker damping material...

If you've been building speakers for some time, you know how much guesswork goes with speaker damping and stuffing. The choices seem endless: fiberglass, wool, Dacron, flat foam, convoluted foam, felt, tar, plus various "magic" compounds that you're invited to brush or pour into your new cabinets. Everyone has their own recipe, and who knows if it's a recipe for disaster? Or what effects the vapors emitted by these chemicals might have on the glues that bond your woofer surround to its cone and chassis? In this era of costly, space-age drivers and computer-assisted design, we think such risks are

totally unacceptable. So we went to work to find the ideal solution.

The problems are fairly well-known: a driver transforms electrical energy into mechanical energy. This mechanical energy is transformed into acoustical energy which is radiated to the outside of the cabinet - the useful front wave - and to the inside - the sometimesuseful back wave. Unfortunately, it is also transmitted though the frame of the driver to the cabinet itself, which acts as a very large "cone" of very small excursion. This means that the spurious resonances and vibrations of the cabinet have to be controlled in a predictable and reproduceable way. That's how we came to BLACK HOLE 5 and the BLACK HOLE PAD.

First, THE PAD. It's a thin (1/16 inch) black flexible viscoelastic damping material (filled vinyl copolymer) with maximum performance between 50 and 100 degrees F (we hope that that covers the temperature range of your listening room) and excellent flame resistance - it meets UL94 V-O. Thanks to its outstanding damping characteristics, THE PAD will dramatically reduce the vibration energy stored in the walls to which it is applied.

Easy to cut and apply, THE PAD has a pressure-sensitive adhesive back: simply peel off the release paper and press hard onto a clean surface. You can use THE PAD on just about anything you suspect of vibrating: driver frames, thin panels like car doors, and, of course, the walls of your speaker cabinets. And it can be used to recess a driver without using a router: just laminate enough layers to match the thickness of the driver frame and apply to the front baffle. Finally, it is the ideal material for "constrained layer" wall construction, where two panels are laminated on each side of a damping material for optimum transmission loss. Because THE PAD has a fine grain leather finish, you can wrap an entire cabinet exterior and give it an attractive appearance at the same time!

For applications which require **maximum damping, isolation and absorption,** we've developed BLACK HOLE 5. One and 3/8" thick, BLACK HOLE 5 is a high-loss laminate that provides optimum acoustical damping performance. It consists of five layers:

Thin diamond-pattern embossing, densified with a polyurethane film surface. This unique surface layer dramatically improves the performance of the whole acoustical system, especially the lower mid-range and mid-bass frequencies where simple acoustical foam loses its effectiveness.

One-inch deep polyester urethane foam, structurally optimized for acoustical damping. Highly effective at "soaking" maximum sound energy with minimum thickness.

Barrier septum, 1/8 inch thick. Made of limp flexible vinyl copolymer loaded with non-lead inorganic fillers, it is a "dead wall" that isolates the vibrations in the walls of your cabinet from the vibrations created inside the enclosure. Polyester urethane flexible open-cell foam, 1/4 inch thick. Thanks to special vibration-isolation characteristics, it

decouples the vibrating structure (the wall) from the rest of the damping system, thus optimizing performance. High-loss vibration damping material, same as The Pad. It is strongly bonded to the cabinet wall with pressure sensitive adhesive.



These layers are laminated using an adhesive-free mechanical and thermal process, thus optimizing performance and eliminating the risk of solvent fume damage. BLACK HOLE 5 can be used in any enclosure, as well as for acoustical panels to improve the characteristics of your listening room. YOU PROVIDE THE MUSIC; BLACK HOLE FIVE WILL TAKE CARE OF THE NOISE!



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#### Continued from page 34 THE SUBWOOFER

I solved the problem of the missing lower bass with a subwoofer consisting of a strong, wellbraced wooden box of 3 ft.<sup>2</sup> capacity and a 1950s 12" Wolverine driver that I purchased for \$8 at a church thrift shop. It's one of those die-cast monsters with a stiff paper cone and a large, porous dust cap. I modified it in much the same way as advocated by Brian Smith ("Adjusting Woofers For High Performance," SB 6/89, p. 22), coating the cone with polyurethane and applying felt to the rear of the driver basket. The subwoofer is fed through a hand-wound coil of wire salvaged from a discarded washing machine motor and produces sound below approximately 240Hz. Friend and foe alike are amazed that the Simplexes seem to put out clean bass down to 30Hz.

My second set of Simplexes, which I built for one of our married children, is supplemented by a subwoofer based upon Eldon Sutphin's findings ("Measuring Drivers With Radar," *SB* 3/91, p. 10). It uses a 10" Radio Shack model 40-8053 driver ( $F_S = 27$ ,  $Q_{TS} =$ 0.41,  $V_{AS} = 4.8$ ) installed in a 2.8 ft.<sup>2</sup> vented box. The 21/2-inch-long duct has a 2" diameter.

Quite a number of interesting designs are offered in back issues of *SB*, so take your pick. Powering the subwoofer with a separate am-

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FIGURE 4: Plan and dimensions for the Simplex enclosure.

plifier is another way to improve your sound system. Purchase a secondhand unit and feed its auxiliary input with the stereo signal taken from the headphone jack on your integrated amplifier or receiver.

## PLACEMENT

These speakers are tolerant with regard to room volume. Our children's living room has a sloped ceiling that is 18' high at one end, and the Simplexes work well there, as well as in small- or medium- sized rooms. The reason for this tolerance is the speaker efficiency: I drive them with a 35W/channel Yamaha amplifier, and they have power to spare.

Locate the speakers at approximately equal distances from the listening position and separate them by 7–8'. Place the tweeters at or slightly below ear level for a seated person and at least 3-4' from the room's side walls. Experiment with their distance from the back wall; a distance of 11/2-21/4' usually provides good results. In most cases, people will be seated close to or against the opposite wall, so give the speakers a slight toe-in that best suits your needs.

The smoothest response and best coupling to the Simplexes is often obtained by locating the subwoofers at substantially unequal distances from the nearest three room surfaces. Mine are offset from center stage in one case and firing into a corner in the other. It pays to experiment with subwoofer placement to avoid cancellation or enhancement of the deep bass.

I also installed a broadband absorption medium behind the speakers. It consists of a 1/2inch-thick fiber mat behind heavy woolen drapes and extends across the entire width of my basement listening room. By absorbing 50– 75% of all sounds above 500Hz, it effectively improves both stereo imagery and ambience.

## PERFORMANCE

The Simplexes are clean and airy, and you can listen to them for hours without fatigue. They evoke an intimate presence, and the imaging is precise. As for timbre, string ensembles sound natural-the cellos full and firm, the violins resinous yet silky smooth. The inner voicing evident in combos makes it seem as though they are right in the room. Woodwinds are delightful, and brass are bright and sharp but not shrill. Symphonic music is spacious, with no loss of clarity or detail in the crescendos. Triangles shimmer and bells twinkle naturally. The piano comes across convincingly, provided you have the subwoofer tuned right to handle the bottom end. Voices, male or female, are pleasurable, whether solo or choir.

Like any good speaker, the Simplex is merciless with regard to turntable rumble, surface noise, mistracking, distortion, and tape or FM hiss. It also will expose goofs. This may spoil listening to some of your favorites, but it will make you nearly ecstatic when auditioning the high-quality recordings you own.
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# LIVING WITH A SPEAKER BUILDER: A VIEW FROM THE OTHER SIDE

By Nancy MacArthur

Living with a speaker builder is not always easy. I don't worry about the noise, the mess, or the smell of drying glue. What bothers me is the complete change of personality—a Dr. JekyII/Mr. Speaker Builder transformation—that my husband Duncan undergoes whenever he starts designing a new set of speakers.

Under normal circumstances I like living with Duncan. His face lights up when he sees his family at the end of the day, and he's handy to have around the house: he can fix anything and does more than his share of housework. But "normal" is not a word I would apply to anyone in the throes of a Speaker Building Fit.

### COMMUNICATION BREAKDOWN

Some stages of the Fit are relatively easy to cope with. The only inconvenience resulting from the construction of speaker boxes, for example, is an occasional unspeakable buzzing sound emanating from the garage. During this stage, I can still communicate with Duncan if only via hand signals. He can't hear me, because he's wearing earplugs and goggles as he hunches over a roaring table saw. On the other hand, if he sees me waving at him frantically from the door, he's willing to take a break to find out what's going on.

Communication is infinitely more difficult during the earlier stages of the Fit, when Duncan starts to design a new speaker. He secludes himself in an out-of-the-way corner of the house, surrounded by piles of circuit

### ABOUT THE AUTHOR

Nancy MacArthur is a freelance writer who lives in northern New Mexico with her husband Duncan, a seasoned audio fanatic, and their small son Colin. diagrams and old issues of SB. Staring off into space, he murmurs such cryptic comments as: "A five ohm resistor and a forty microfarad capacitor will equalize the woofer's impedance," or "Low WAF on this one."

He won't tell me what "WAF" means, but it must be something truly dreadful. Once the Fit has passed, he's usually only too happy to explain to me, in detail and at great length, why a resistor is different from a capacitor, and why he can't live without the new ones advertised in the latest catalog instead of using up the mounds of tiny parts he has piled on his desk.

I've thought about taking a laissez-faire approach to Duncan whenever he starts designing a new set of speakers. He could stay hunkered down in his corner, muttering into his catalogs and parts lists. I could go on with my regular activities, pausing only to toss an occasional haunch of meat his way. He would avoid starvation, and I would avoid getting irritated with him. The trouble with this approach is that sooner or later a problem will arise which I can't solve without his input.

The usual methods of gaining his attention don't work when he's in the design stage of a Speaker Building Fit. For example, under normal circumstances, I can count on an enthralled response whenever I mention the word "food." But during the extreme throes of the Fit, food is of much less interest to him than the visions of new speakers floating over his desk.

Getting through to Duncan during this phase requires combining the persistence of a New York street vendor with a creative disregard for the truth. I start with a straightforward approach:

"Duncan, I need to talk to you."

"Mmph."

"Supper's ready," I add craftily.

"Jussec." He burrows deeper into his piles of circuit diagrams.

"Nuclear war was just declared."

"Arglemmph."

"Duncan, the baby just set fire to your power amp."

"Mmmm...What did you say?"

It's simply a matter of understanding his priorities.

#### FIT OF PASSION

If I'd had the sense God gave a bunny rabbit, I would have seen what was in store for me years ago on our first date, when he spent two hours lecturing me on why I should build a transmission line as opposed to a bass reflex speaker. Luckily for Duncan, I already had a crush on him or sixteen years of happy marriage might never have come to pass.

l even get into trouble when I try to help him. One holiday season a few years ago, Duncan decided that the only present he would consider was—you guessed it—raw drivers for building new speakers. I called a well-known speaker company, ordered the drivers, and asked if they would arrive in time for my husband to open at Christmas. At first, all I heard on the other end of the line was stunned silence.

"Hello? Hello?"

"What do you mean, you want to give your husband speakers for Christmas?" a deep voice bellowed back at me.

I was bewildered. "Well, you do sell them, don't you?"

A weary sigh came over the line. "Look, lady, whenever someone's wife calls me, it's because her husband is spending all his time in the garage building speakers and she wants to yell at *me* for selling him the drivers. What are you, some kind of nut?"

l guess so, Mr. Speaker Man. l guess so. 🔈

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Reader Service #31

# THE IMP GOES MLS

# **By Bill Waslo**

I MP is an inexpensive circuit board and software complement which makes a PC into a sophisticated FFT analyzer for loudspeaker and audio equipment evaluation and development. I introduced it to the world in SB 1/93 (p. 10). When used for measuring transfer functions or impedances, IMP operation is based on the fact that frequency response is the Fourier Transform of a time impulse response. This mathematically sound concept yields very accurate results and close agreement with measurements obtained by other means.<sup>1</sup>

A narrow pulse is attractive as a measurement stimulus for several reasons. It is easy to generate using inexpensive circuitry. It contains a wide frequency spectrum allowing simultaneous measurement of most or all of a speaker's range. Both the phase and magnitude of its spectrum are essentially uniform over a wide range of frequencies. In the time domain, echoes in a device's pulse response are easily identified and removed.

#### **NOISE INTRUSION**

Pulse testing's weakness, however, is the test signal's rather low energy content.<sup>2</sup> The pulse stimulus is very brief, while the response data collection time must be many times longer in order to provide low-frequency information. This affords an opportunity for noise to intrude into the measurement.

The test signal energy cannot be arbitrarily increased by "turning up the volume" into the unit under test. The crest factor (defined as the ratio of the signal's peak power to its average power) is very high for a pulse stimulus. You may clip your amplifier or drive your speaker into nonlinear operation with many watts of peak power, while only applying microwatts of average power to do battle with the noise. Due to the nature of most environmental noise and the need for long response collection times, the problem is most severe at lower frequencies. In the original IMP, noise is dealt with by employing repetitive pulse stimuli and averaging. The stimuli must be spaced enough in time to permit obtaining low-frequency information and allow device pulse responses to decay sufficiently. For every doubling of the number of time responses averaged, the signal-to-noise ratio, neglecting quantization and coherent noise, theoretically should increase by 3dB.<sup>3</sup> The trade-off is measurement time versus increased noise immunity. Averaging two responses yields 3dB improvement, 6dB with four responses, and 9dB with eight.

The simple averaging scheme works quite well for most speaker measurement purposes,











FIGURE 2: IMP/MLS satellite single-sided PC board pattern.

and for home builders who can afford the extra time and have reasonably quiet surroundings. If you wish to increase the noise rejection by 30dB, however, you'll need to average over one thousand responses. That might take longer than you care to wait.

Other broadband stimuli are also characterized by flat spectra magnitudes, but possess friendlier crest factors than the pulse. One is the frequency sweep or its optimized version, the chirp. This stimulus is not as simply generated, and phase information can be difficult to obtain unless the system being measured is already definitely known to be minimum phase. Random white or pink noise is another broadband test signal in which the magnitude spectrum is generally flat or gradually sloped but the phase is random. Due to this random factor, two channels (input and output) must

FIGURE 3: Component placement guide.

be measured simultaneously for accurate determination of a system response.

#### **PRECISION NOISE**

A very convenient stimulus is pseudorandom noise, which is an analog version of a digital signal known as a Pseudorandom Number (PN) pattern or Maximum Length Sequence (MLS). PN sequences are used extensively in spreadspectrum radio communications, data encryption, music synthesizers, and even computer games. In pseudorandom noise, the magnitude of the spectrum is basically flat, while the phase is scrambled—but not really random. The spectrum is absolutely deterministic and repeatable, like that of the pulse, so only a single measurement channel is required.

The MLS has the additional property that its autocorrelation function yields an impulse

signal, and the cross-correlation function of a system's response to an MLS with the MLS itself is that system's impulse response. An example might help clarify the meaning of this intimidating-sounding statement.

A maximum sequence length of seven, modified so that digital zeroes are represented as negative ones, is the series:



After the last value, the sequence returns to the beginning and repeats. If one copy of this sequence is lined up beneath another, and the corresponding values are multiplied and their products summed, the resulting value is seven:



Due to the MLS's periodic nature, a circular shift of digits to the left corresponds to a time delay (any digits that run off the left can be pasted back onto the right). If the same multiply-and-sum operation—a form of correlation—is performed instead with a timeshifted MLS as the lower sequence, the result will be the value minus one. For example, here's a shift to the left of five places:



A plot of the autocorrelation result versus the time shift shows a large peak at zero shift, and at multiples of seven, and a small negative value elsewhere:



This plot is quite similar to a periodic impulse signal. If we used an MLS with a length of 4,095, the peak value would be 4,095; it would repeat every 4,095 points, and the "baseline" value would still be only -1.

#### **CROSS-CORRELATION**

Suppose that an N-point MLS were converted into an analog signal by making each value a sample into a digital-to-analog (D/A) converter, and then fed repeatedly through a sys-

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tem to be tested. Suppose you then digitize N points of the resulting response from the system, and perform the multiply-and-sum operation of this data stream versus that of the original MLS stream for every circular shift (called a cross-correlation operation). The result versus time shift plot would be the same as the system's time impulse response—or very nearly so, assuming that the impulse response decayed sufficiently within the time period of the N samples.

Such is the idea behind MLS testing. Several cycles of the N-length sequence are fed into the unit being tested, and its response is subjected to some intensive number crunching. The resulting impulse response is the same as you would get from an IMP-type stimulus, but with an important difference. The benefit of adding this more complex stimulus and all of the computational overhead is that the resulting impulse response has the noise immunity of averaging the responses to N-actual pulses. In addition, any unintentional transient noise picked up during the measurement is spread evenly across the time interval by the cross-correlation operation and appears as more benign low-background noise. All of this is achieved within the time period of a single data collection and two N-length sequences, allowing for one "warm-up" run. When N is equal to 4,095 or more, the improvement in noise immunity and reduction in measurement time are considerable.

The technique dates back to 1979: Around 1982, the Fast Hadamard Transform was applied to greatly speed up the correlation arithmetic.<sup>4</sup> Commercially, MLS stimulus testing has been made available in the DRA Laboratories' MLSSA system, which allows use of sequence lengths from 4,095 to 65,535 points.

#### **IMP SATELLITE**

I have designed an add-on board to the IMP main board. When operated from the IMP software version 2.0, it will allow 4,095-point, MLS-based measurements to be performed. The software utilizes an integer-based version of the Fast Hadamard Transform, which makes clean, quiet impulse response acquisitions extremely quickly—typically, before you are ready for your next step.

The IMP/MLS satellite board schematic is shown in *Fig. 1*, with the parts list in *Table 1*. The MLS sequence and control signal are stored on CMOS programmable read-only memory (PROM) to minimize required hardware. The original IMP pulse stimulus is also coded onto the PROM to allow you to choose either stimulus via software control.

Two other ICs address the PROM and latch the data and control signal. The output latch is configured with weighted summing resistors for use as a seven-bit D/A converter to provide inexpensively for possible future enhancements. The latch IC supply lines are filtered to prevent digital noise feedthrough. Note that with this arrangement the latch IC gets its power supply ground from analog ground—be sure this ground is connected. *Figures 2* and 3 show the satellite's single-sided PC board pattern and component placement guide, respectively.

Continued on page 44





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PHOTO 1: Recommended mounting arrangement for MLS board.

#### Continued from page 42

Six single wires and two twisted pairs are run between the IMP/MLS and IMP boards for signal and power exchange. In addition to connecting wires from the satellite board to pads on the IMP board, you must remove resistor R23 from the main board and cut one trace on the IMP board's solder side (*Fig. 4*). Use a sharp razor knife to cut and remove a small section of the indicated trace. To build the satellite board, first solder the IC sockets and all passive components (don't forget jumper JMP1), then decide where the board will be mounted. Avoid placing it too close to the analog circuitry near the front of the IMP board, or you may end up generating rather than suppressing noise problems. The satellite board is designed to mount piggyback style over the back of the IMP board via two mounting holes and standoffs (*Photo 1*).

After you have determined the mounting location, cut the trace as shown in *Fig. 4* and the solder wires as follows—no lengthier than necessary—from the IMP/MLS board to locations on the solder side of the IMP board. Connections to the main board must be surface-soldered on existing pads. The lengths in parentheses are appropriate for piggyback mounting.

- +5V and GND (color-coded twisted pair) of the MLS board to the positive and negative pads, respectively, of C4 on the main IMP board (4.75")
- Test Out and AGND (another color-coded twisted pair) from the satellite to the center conductor and shield of connector J4 (6.25")
- Point CC of the MLS board to U11 Pin 8 on the main board (4.75")

Continued on page 46











FIGURE 7: MLS-derived loudspeaker response.



FIGURE 9: Waterfall plot of Cal signal, single MLS stimulus.



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Continued from page 44

- Point I/M\* of the MLS satellite to the main board pad corresponding to J5 Pin 4 in *Fig.* 4 (4.75")
- The pad marked A1 on the MLS board to U15 Pin 9 on the main board (4")
- Point GO\* on the MLS board to U14 Pin 6 on the main board (3")
- Point T of the MLS board to U12 Pin 5 on the main board (4.25")
- The pad marked HBEN on the satellite to U3 Pin 3 of IMP board (4.75")

Install ICs U1, U2, and U3 into their sockets. Notice that when the satellite is to be mounted as shown in *Photo 1*, the satellite board ICs mount in the opposite orientation to that of the main board ICs. Remount the IMP board, and then mount the satellite board. You should now be ready to go.

#### **OPERATION**

In order to access the modified IMP's MLS capabilities, you must use the new Version "M" IMP software. Under the [Acquire]

menu, an option called "MLS" now toggles between off and on when selected.

You can see what the MLS signal looks like in the time domain by arranging a typical IMP measurement, setting MLS on, and selecting [Acquire Repeat]. As with the pulse stimulus, set the appropriate input gain control so the trace peaks are within the IMP input's linear range (i.e., within the display window). You will now have a continuous stream of what looks like noise rather than a single peak (*Fig. 5*). If you use [Esc] to stop the stimulus and vary SIZE—using F1 after each change to redraw—you can observe the MLS signal on various scales.

Selecting [Acquire Collect] will result in a different picture. The Fast Hadamard Transform will now be applied between the data acquisition and the display. The resulting trace will resemble the old familiar IMP signal, and, from this point on, can be processed in the same ways but will yield greatly improved noise rejection.

An important point: the input gain adjust-

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ments must be made for the MLS signal peaks only as you see them when using [Acquire Repeat]. The size of the pulse peak after using [Acquire Collect] is not directly related to the IMP input dynamic range when in MLS mode. Typically, MLS-derived pulses will be much lower in amplitude than the normal time impulse responses, due to noise peaking when the MLS is passed through a speaker.

If you wish, you may still average responses when using MLS stimulus by using new selections under the [Setup Average\_num] menu. You can also configure the program to start up in MLS mode (as well as at your selection of most other program settings, such as frequency range, screen colors, or response smoothing) each time you use it. This new Version "M" software also features mouse support, user macros, on-screen titles, and overlaid plots. Additional information on using MLS and other software features can be obtained with the new position-sensitive online Help facility, reached via Shift-F1.

### COMPARISONS

The equivalence of pulse-derived and MLS data is illustrated in *Figs. 6* and *7. Figure 6* is a loudspeaker time and frequency response measured using 15 averaged pulse stimuli and removal of the more prominent echoes. The "m" near the IMP logo in the lower plot indicates that Cal has been used to correct for pulse shape and filter characteristics, and that a microphone calibration file also has been factored into the plot to correct for the microphone's response. Microphone calibration data can be provided with the Old Colony Mitey Mike and with test microphones from Josephson Engineering.

*Figure 7* shows equivalent data taken using the MLS stimulus in a single acquisition. I adjusted the GAIN parameter so the 1,005Hz measurement agrees with that of *Fig. 6*. The two plots are virtually identical.

*Figures 8* and 9 are waterfall plots of the Cal signal taken with pulse (five averages) and MLS stimulus (one), respectively. The MLS-derived plot shows a marked advantage in these figures (the scale is 15dB/division).

One situation in which IMP may work

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 Rife, Douglas D. and John Vanderkooy, "Transfer-Function Measurement with Maximum-Length Sequences," JAES (Vol. 37, 6/89): 419.

3. Fincham, L. R., "Refinements in the Impulse Testing of Loudspeakers," JAES (Vol. 33, 3/85): 133.

4. Borish, Jeffrey and James B. Angell, "An Efficient Algorithm for Measuring the Impulse Response Using Pseudorandom Noise," JAES (Vol. 31, 7/83): 478.





better with the pulse rather than the MLS stimulus is when far-field measurements are made in highly reverberant rooms. The pulses can be spaced apart in time intervals of arbitrary length, but the IMP MLS is mathematically constrained to repeat every 4,095 samples. If the time response hasn't sufficiently died out within such a period, the response tail will reappear both before and during the beginning of the recovered pulse response. This is shown in the top plot of Fig. 10, and can be compared with the pulse-derived equivalent in Fig. 11. The sixth-octave smoothed frequency response plots of the two measurements (lower plot in each figure) nonetheless do not seem to differ greatly.

The MLSSA system allows for use of considerably longer MLSs to avoid such potential problems in slow-decay-rate environments. With IMP, you could still use the pulse stimulus in such reverberant rooms, but a few dollars spent on throw rugs or wall treatment would be more to the point and would improve the sound as well. Using the lower sample rate for bass measurements will also stretch out the MLS time period thirty-two fold, usually out of danger's way.

#### **DRIVER PARAMETERS**

The MLS stimulus greatly improves the repeatability of T/S measurements using IMP. In addition, measurements are made much more quickly, room noise is hardly a concern, and added-mass launching is much less likely using the MLS.

Very-low-frequency impedance measurements are more reliable, and even IMP's estimation of the voice coil DC resistance is more trustworthy, reducing the need for an accurate ohmmeter. Best of all, those annoying 60 and 120Hz bumps often caused by line pickup in the cables—50 and 100Hz for some of our overseas IMP users—no longer appear in the impedance plots.

If you find you don't have time for all the IMP measurements you'd like to make, if you don't like asking others to be "quiet for just a minute" while you average another bunch of responses, or if you'd just like to increase the performance of your IMP analyzer system, MLS capability is well worth including. I don't think we'll be hearing much longer about how amateurs are building speakers almost as good as the commercial units. Now that we can provide intensive individual care and also have access to inexpensive measurement gear and design software, I think the speaker manufacturers will soon be comparing their products to those of the amateurs.

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## Test Disc III Hi-Fi News & Record Review

This newest addition to HFNRR's world-renowned test library was produced by Trevor Butler and executive produced by Steve Harris. Includes description/instruction booklet. 1993; from the United Kingdom; one disc; 72:36. HFN #020.

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#### Speaker System Designer 4.0 Software for Windows

SSD4.0 is an extension of the previous SSD3.0 DOS package. It requires Windows 3.1 and enables the designer to create, evaluate, and then optimize 2-, 3-, 4-, or 5-way loudspeaker systems prior to starting enclosure assembly. The designer is also able to model the behavior of the crossover when loaded by the driver in an enclosure and observe the dramatic effect on the frequency response curve of the crossover which the driver may have. A suitable design compensation network can then be created by another SSD4.0 tool and the effect on the crossover re-examined. The program allows the designer to pinpoint problems and test possible solutions so that the frequency response of the crossover is as close to the "ideal" as possible.

Functions available in this package include driver reference library creation; loudspeaker enclosure design and optimization; compensation of the driver impedance or amplitude; crossover filter design and optimization; system frequency response evaluation and optimization; frequency response of the system "in room"; L-pad, series LRC, and zobel network calculators; impedance peak suppressor; and thermal analysis, which enables the user to examine the frequency response and structure temperatures at all input power levels.

Environment: Windows 3.1. User interface: mouse, keyboard, Windows. Program control: mouse, menus, TAB key for data entry in dialog boxes, cursor keys for data entry in display windows. Mass storage: SAVE and LOAD menu options. Hard copy: graphics printer, preferably with banding capabilities. Screen output: VGA, 640 × 480 pixels, 16 colors, or SVGA, 800 × 600 pixels, 16 colors. Processor: 286 minimum. RAM: 2Mb minimum. Hard disk required: 2.5Mb minimum plus 1Mb to install. By Bodzio. Detailed instruction manual included. Purchasing options available:

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# Wayland's Wood World

# LET'S GET THIS STRAIGHT By Bob Wayland



PHOTO 1: Making a test line with a marking knife.



**PHOTO 2:** Results of the flopping interface for a try square. Chalk is used to make the scribes visible.

Perhaps one of the most frustrating aspects of building speaker enclosures is ensuring they are square. After carefully cutting out each piece and preparing the joints, we all too often become careless in gluing up the carcass. The resulting slightly off-square box makes fitting the back, front grille, and stand difficult. The greatest discomfort is seeing the Leaning Tower of Pisa every time you look at your speaker. You can purchase very expensive, highly accurate squares which are a minimal help, but then you must be very careful never to drop or in any way abuse the delicate tool. All of this is inhibiting, but, thankfully, there is a simple way around the complications.

Woodworking, like all activities, has its

share of tricks that extend throughout the trade. When checking whether something is square (a cut, a try square, or what have you), the trick is the *flopped interface*. If a right angle is slightly off, and you separate along the "perpendicular" line, turn one part over by 180°, then replace it next to the other part with the bases remaining on a straight line, any



PHOTO 3: Peening the blade of the try square to obtain alignment.50 Speaker Builder / 6/93



**PHOTO 4:** A 2 × 4 crosscut with a 1° error. The saw table is used as the reference flat surface.

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PHOTO 5: A 2 × 4 crosscut with only a 0.1° error.

**PHOTO 6:** Saw blade check on a piece of  $2 \times 2$  stock.

variance from perpendicular will be amplified by two. With care, you should be able to detect an error on the order of  $V_{20}$  of a degree.

#### **SQUARE OFF?**

The first step is to be certain that your tools hand and power—are set up square. One of the handiest testing tools you can own is the try square, although those you buy at the corner hardware store tend not to be true. To determine whether it is, find a piece of scrap wood with a straight edge. Be certain to test the straightness against a good metal straightedge or flat surface, such as your saw's tabletop. Place the square on the straight edge and mark a line on the wood along the perpendicular blade. We are concerned with accuracy, so use a marking knife (*Photo 1*). This provides a narrower line and will also be much closer to the blade.

Flop the square to see if the blade aligns with the line. When I tried this with an old, cheap try square, the result was the first set of lines in *Photo 2*. To avoid replacing the try square, I decided to straighten it by peening the metal blade at the corner with a sharp pointed punch. Do this either at the top or bottom of the blade, depending upon which way you want it to go (*Photo 3*). As you can see, this is usually a two-step process. You could have filed the blade to obtain alignment, but you would have only one side perpendicular unless you filed both edges. Besides, that is a lot of unnecessary work.

Before worrying about the squareness of your enclosure, you must start with square pieces. Once again, the accuracy of your saw's miter gauge and the perpendicularity of the saw blade are in question. We can use the same technique as described above to ensure an accurately aligned table saw. A 1° error in

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PHOTO 7: A suggested order for squaring up stock.



PHOTO 8: Checking for squareness with a try square.

the miter gauge will produce unacceptable results in even a common  $2 \times 4$  (*Photo 4*).

For this technique to work, the top and bottom faces must be exactly parallel and the comparison made on a flat reference surface. The same applies for an error as small as  $0.1^\circ$ , as seen in *Photo 5*. While this is over a short distance of 3.5'', you normally will be working with widths of ten or more inches—so

watch out. The owner's manual for your saw should provide detailed instructions on how to align it.

The check for the saw blade relative to the saw table (*Photo 6*) was made with a  $2'' \times 2''$  board. The blade in the left-hand cut was  $\frac{1}{6}^{\circ}$  off perpendicular and dead on for the right-hand cut. The two pieces on the right have been set with a small angle between them so

you can spot any deviation more easily. For an idea of this adjustment's importance, think about a  $1/6^{\circ}$  error in a butt joint at the end of a 20" board.

Let's consider for a moment the best way to square up a side (or, for that matter, the top/bottom or front/back) of your enclosure. Begin by creating a straight edge that is perpendicular to the face of the board, preferably



World Radio History



PHOTO 9: Pointed edge of the squaring rod.



PHOTO 10: Using squaring rods to measure a diagonal.

in the direction of the grain if you are working with solid wood (*Photo 7*). This task can most easily be done on a joiner, but you can also use a table saw followed by a careful pass of a joining plane.

With this reference edge, use your table saw to cut a parallel edge at the desired width. Then make the cross cuts to cut the side to length. Each step is very dependent upon your saw's accurate alignment. You can make the same cuts with a hand or circular saw, but this requires your undivided attention.

#### **DIAGONAL DIAGNOSIS**

Once all the pieces are true and square, you are ready to glue up your enclosure. Perhaps the most common mistake is to use a try square to check for the alignment (*Photo 8*).

While this provides a useful first approximation, it is unsatisfactory if you check only one corner. On assemblies with long sides, it can gauge only a fraction of the lengths involved. If the slightest curve or taper exists, or if the sides aren't straight across their bottom edges, a try square won't work at all.

A more accurate (and easier) technique is to ensure that the diagonals of the enclosure



Reader Service #11



PHOTO 11: Checking the opposite diagonal.



PHOTO 12: Using the enclosure back to square up.



are the same. You will first need to make a pair of squaring rods, sometimes called pinch rods. Choose a stable, straight-grained wood for a pair of rods about  $3/4'' \times 1/2''$ , say three-quarters the length of the diagonal. A fter you have built a few enclosures, you will have your own collection of various lengths. Put a tapered edge along the wide dimension at one end of the rod, somewhat less than 90° (*Photo 9*), to allow the edge to easily fit in the corner of your enclosure.

Place the rods across a diagonal as shown in *Photo 10*, and clamp them together with two small C-clamps; then test the opposite diagonal (*Photo 11*). If they are the same, your enclosure is square; you can make corrections by tilting the clamps. With a difference of, 1/8'', for example, the carcass will be 1/16'' out of square. The size of your enclosure will determine whether you are willing to live with this: the larger the enclosure, the more tolerable the slope.

I often use another simpler method. Carefully cut the back for your enclosure, being absolutely certain that it is square. Loosely clamp up the enclosure and place the back into its opening (*Photo 12*). Choose one corner and screw it into the backing brace. When you tighten the clamps, the enclosure will automatically be square. Take care that you don't glue the back to your enclosure.

If you have questions, please send them to me via *SB* and I will try to answer them in future columns.

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# Tools, Tips & Techniques

# FINDING RESONANT FREQUENCY

Finding the point of zero phase shift is the best way to determine resonant frequency. My second-hand scope doesn't do Lissajous traces; however, it has two inputs with inversion switches and a "chop" switch which causes the superimposed display of both channels.

I measure impedance using the constantcurrent method, which puts about a  $700\Omega$ resistor in series between the signal and the driver, by attaching one set of input leads (with the invert switch for that channel off) to the signal generator, and the other channel (with the invert switch on) to the load (driver). If both channels are carefully zeroed, the resonance is the frequency at which the two traces cross each other and the zero line at the same point.

Don Stauffer Dallas, TX 75211

# LAZY BEARINGS

If you have large, heavy speakers, and have them placed on flooring other than carpeting, experimenting with toe-in can be a difficult proposition. Rotating the speakers can cause damage to both cabinets and floors, while lifting them can be a job for two or more



PHOTO 1: "Lazy Susan" bearing.

people. One workable solution is to mount them on "lazy Susan" bearings (*Photo 1*), several ball bearings in a raceway between two pieces of stamped sheet metal, which permit the metal pieces to turn easily with respect to one another even with a heavy load imposed on them.

The bearings can be fastened to the bottom of the speaker cabinets with double-sided adhesive tape. Unwind a short length of tape from the spool and peel off the backing to expose both adhesive-coated sides, then simply apply light pressure to stick the tape to the surface.

Placing the bearings between the cabinets and the floor allows you to turn the speakers easily. Several drawbacks do exist: the speakers may turn too easily, making it difficult to maintain the desired toe-in, and the bearings may have too much play and wobble. Both problems are solved by adding a small wedge, marble, cone, or large machine nut between



the cabinet bottoms and the floor to lock the speakers in place and add stability.

These bearings can be a big help when setting up speakers and later removed, if you desire, when you have determined the ideal toe-in.

James T. Frane Orinda, CA 94563

# ACCURATE MEASUREMENTS

I have always had trouble taking the accurate measurements necessary for vented-box design, especially at frequencies below my DVM's rating. Readings were tedious, inaccurate and inconsistent from one sitting to the next.

I discussed this problem with my friend, John Levreault, who came up with an oscilloscope method which not only solved the problem of accurate impedance measurement, but also protected readings from dial inaccuracies on my audio signal generator. I can't believe how much this has helped. It was also less tedious.

Instead of reading frequency in hertz (and impedance in ohms) as absolute numbers, read them as *relative* oscilloscope scale readings wherever possible (when the equations in which the values are used are proportions in which the units cancel). For example, to find the  $Q_{MS}$ ,  $Q_{ES}$  and  $Q_{TS}$  of a woofer find the resonant frequency and set your scope time base so a complete half-cycle will show at the F1 frequency (which must be found using the DVM). This will take some experimentation; usually two or three half-cycles will show at the resonant frequency.

Now count the number of time units for all the complete half-cycles showing on the scope at the resonant frequency, and write down the proportion of half-cycles to time units. Don't touch the time base knob again after this.

Read F1 and F2 the same way as  $F_S$  and use the value of proportions in the formulas for  $Q_{MS}$  and  $Q_{ES}$ , as well as the double check value,  $\sqrt{F1 \times F2}$  (which should be near  $F_S$  if your measurements are accurate). The results will be mathematically the same as if you had used the absolute frequencies.

John sets his signal generator extremely low to measure  $R_E$ , allowing him to even measure  $R_{MAX}$  (or  $R_M$ ) and  $R_E$  as relative values to calculate  $R_O$ . My generator doesn't go that low, so for me  $R_{MAX}$  will have to be an absolute measurement, as will  $R_E$ . I highly recommend the Wheatstone Bridge ("A Wheatstone Bridge for Your Voice Coil," *SB* 5/89, p. 36) to measure voice coil resistance; this is perhaps the most mathematically sensitive value of the whole process, and should be as precise as possible ( $0.05\Omega$  accuracy is not unreasonable). I use the constant-current method and a good DVM for R<sub>MAX</sub>, which is OK because usually F<sub>S</sub> isn't too low for my DVM to be stable.

What you are really doing is reading absolute frequency (or impedance), but not in hertz or ohms. As long as you keep the time base (or, if you're measuring impedance using the constant-current method, the voltage scale) constant, the results will be correct in any formula where the units cancel.

This also works well when measuring

woofer compliance, or analyzing whether a box is on target (refer to Bullock's "Fine Points of Vented Speaker Design," *SB* 2/81, p. 18). I have used this technique with the following formulas:

For driver parameters:

$$Q_{MS} = F_S \times \frac{R_O}{(F2 - F1)}$$

$$V_{AS} = V_{BT} \times \frac{(F_{H}^{2} - F_{M}^{2}) \times (F_{M}^{2} - F_{1}^{2})}{(F_{H}^{2} \times F_{1}^{2})}$$

For Q adjustment (SB 2/81, p. 22), FSB





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itself can't be found using this method, since its formula has two frequencies in the numerator but only one in the denominator:

$$\mathsf{F}_{SB} = \frac{(\mathsf{F1} \times \mathsf{F}_{H})}{\mathsf{F}_{M}}$$

However, FSB is used in the following formulas for modified O values:

$$Q_{TB} = \frac{F_S}{F_{SB}} \times Q_T$$

$$Q_{MSB} = \frac{F_S}{F_{SB}} \times Q_{MS}$$

$$Q_{ESB} = \frac{F_S}{F_{SB}} \times Q_{ES}$$

Since F<sub>S</sub>/F<sub>SB</sub> is always used as an adjustment factor to the Q values, we can find:

Adjustment Factor = 
$$\frac{(F_S \times F_M)}{(F_H \times F_1)}$$

using this method instead of Fs/FsB to adjust the QT, QMS and QES.

For target analysis:

$$H_{A} = \frac{F_{M}^{2}}{(F1 \times F_{H})}$$

$$\alpha_{A} = \frac{(F_{H}^{2} - F_{M}^{2}) \times (F_{M}^{2} - F_{1}^{2})}{(F_{H}^{2} \times F_{1}^{2})}$$

Don Stauffer Dallas, TX 75211

# **NEEDLING CONES**

For anyone whose speaker surround is peeling off a plastic cone, I have a solution. Several years ago I bought a case of Seymour Sound woofers which employed a plastic surround and slippery plastic cone. Over time they had delaminated, and were repaired with rubber cement. Recently, when trying to use my last driver in a dual-vented bandpass system, the cone repeatedly fell apart. The pressures were too much for the fragile bond.

Finally, I glued and sewed the cone to the surround. I drilled a series of holes about 1/4" apart along the circumference through both the cone and the surround, using a Dremel tool with a PC-board bit. I then stitched the cone to the surround with light fishing line.

For best results, use two needles and stitch from both sides as you would in leather crafts. Seal the stitches and any air leaks with a good cement, like Pliobond or RTV silicone. Your system will withstand the pressures of time.

Matthew Honnert Carol Stream, IL 60188

# **FINISHING TOUCH**

Since I started reading *SB* three years ago, I've noticed a lot of attention is paid to cabinet construction, and almost none to different types of finishes. Let's face it: the speaker may sound wonderful, but if the finish isn't high-caliber, you'll be reluctant to put it into your living room. I have tried various methods of finishing speakers, such as staining, lacquering, and so on.

On my last project, Ralph Gonzalez's excellent Delac-10s, I used Multispec Fine Speck<sup>TM</sup> paint. A gallon will generally be enough to cover two to three pairs of small-to-medium-size cabinets. Fleckstone<sup>TM</sup> should yield similar results. Both of these paints produce a granite-type finish which requires little preparation, and are available in a variety of colors.

You will need an airless spray gun to apply the Multispec (you can probably use a compresser-type sprayer, as well). Considering what we spend in the pursuit of better sound, the price of the sprayer is not too much, besides having other uses around the home.

Before applying the paint, fill in any gaps, screw holes or large gouges with wood filler. The cabinets then just need a rough sanding with 100–120 grit paper. No need for a mirror-smooth surface with this stuff, as it covers just about anything.

The rate at which you apply the paint and the diameter of your spray tip will determine how the paint grain will look. Most spray guns have a flow regulator you can adjust for applying different types. Experiment to find the setting that gives you the best look. I applied my paint with a Krebs model 40T, which tends to make the grain look less pronounced, while the Graco produces thicker, more marbled grain lines.

Avoid working too close, because the paint will run; however, standing too far away will yield a poor grain pattern. As with most types of spray painting, you should remain 10-12'' from your project, use nice smooth strokes, and avoid overspraying. The paint usually dries in one to two hours, but I leave mine overnight before doing any touch-up. I like this paint for its ability to cover most jobs in one coat.

Read all the directions before you begin painting (which most of us, myself included, never do). Do not mix the paint too vigorously: you are just stirring up the heavier particles from the bottom. Overstirring will mix the different colors instead of leaving them suspended in the paint, thereby killing your granite look. This paint does not like cool temperatures, so stay within the manufacturer's guidelines. Once the paint is dry, you can obtain a polished look by spraying or brushing on a few coats of water-based high-gloss polyurethane, and then sand lightly with fine-to-ultra-fine sandpaper between coats. This also produces a harder, more durable finish. My work with Multispec has always been on MDF, although I did spray a test sample of some pine and it had much the same appearance. So your results shouldn't vary much from one type of wood to another. This look may not be for everyone, but it *is* different, aesthetically pleasing, and fairly easy to use. In my opinion, this adds up to more time spent listening to my latest project instead of sanding smooth that 18th coat of lacquer.

#### Gary Riolo The Woodlands, TX 77381



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# Craftsman's Corner

# A HORN FOR THE GYM



PHOTO 1: Three-way horn for gymnasium PA system.

I wrote to Bruce Edgar to ask his advice on the suitability of his 70Hz corner horn design for use in a gymnasium PA system, and his reply was most helpful ("Horns Aplenty," *SB* 4/93, p. 76). I subsequently built two systems (*Photo 1*).

In my original letter, I underestimated the size of the gym, which is actually  $100' \times 50' \times 24'$ . I built the 70Hz horn, Bruce's midrange horn, and bought the Pyle H2610 horn for the tweeter. The bass driver is the Pyle 6520, and the midrange is the Pyle MH516 stage monitor. I built a 6dB crossover at 500 and 5,000, treating the 4 $\Omega$ woofer as an 8 $\Omega$  due to the horn loading.



The three-way system easily and cleanly fills the entire room to uncomfortably high levels at very low inputs. Typical input levels (according to the meters on the inexpensive receiver the gym uses) are less than a full watt. In the 25W range, kick drums are readily palpable at 30' from the system. The sound is quite clean with no audible distortion at any level, although you quite clearly hear any inadequacies in the input. The sound is also uncolored: on PA, people actually sound like themselves. Further, with only one system operational, a trip around the gym floor disclosed no changes in frequency balance or



Reader Service #38



PHOTO 2: Horns wall-mounted on shelves

timbre; the systems seem uniformly cylindrical in their radiation pattern.

The frequency response graph (*Fig. 1*), as measured with a Radio Shack SPL meter in my garage (on a 30" stand in the corner, both doors open) from a Japan Audio Society test CD, was within  $\pm$ 5dB from 70–12,500, with a dip at 800 that may have been a test artifact. (I couldn't hear it except with test sweep tones.)

To date, we have burned out one tweeter coil, which was easily replaced. Since about 100 children a day have access to the system, it was likely abused, but it may be that a steeper-slope crossover should have been employed.

The chalk dust which coats everything in the gym may also have been a reason for the failure, as the systems lived on the floor next to the exercise springboard for many months. They are now mounted on shelves on the wall



PHOTO 3: Detail of horn mounting.

(*Photo 2*). Since they are not in corners, and were only designed to reach down to 70Hz, deep bass is lacking but adequate for the intended purpose.

Stephen Katz Topeka, KS 66601

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Reader Service #29



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World Radio History



*Heath Nostalgia*, by **Terry Perdue**. Available from Old Colony Sound Lab, PO Box 243, Peterborough, NH 03458-0243 for \$9.95 plus \$3 shipping (USA).

Heathkits stimulated my early interest in electronics. I assembled my first Heathkit, a small power amplifier, around 1960. Most of my test equipment started out as Heathkits, with modifications along the way. Most likely many have grown up with Heathkits and share similar experiences.

In the early days, we assembled electronic gear from kits for a number of reasons. First, providing your own assembly labor was a major economic benefit, with the personal satisfaction derived from building your own equipment equally important. In addition, Heath's designs, detailed instructions, trouble-shooting procedures, and customer support ensured success, as well as an education. Since professional test equipment often was very expensive, kits were an affordable alternative for individuals and small service shops.

Since then, the world has changed. Kits are only of minor interest in the electronics industry. Though there are a few education-oriented kits in the latest catalog, Heath no longer produces mainstream kits.

*Heath Nostalgia* chronicles Heath's growth and decline as the major force in the electronic kit industry. After a brief history of the company (followed by 24 pages of pictures), the author delves into the recollections of several people associated with Heath.

Heath Airplane Company was established by Edward Heath in the early 20th century. Among the wares sold by the company were parts kits for various aircraft assemblies, from which one could assemble an entire light plane. Heath's first mail order catalog was published in 1925. After Heath was killed in a plane crash in 1931, the company was sold and renamed the International Aircraft Corporation: it closed in 1934 due to business difficulties.

Howard Anthony purchased the company's name and remaining assets from the IRS in 1935, and moved the business to Benton Harbor, MI to sell aircraft and accessories. Anthony resurrected Heath's training program to build and fly planes, and added aircraft radio receivers and transmitters to the catalog. Heath's staff grew to 80 employees during World War II.

After the war, business shrank drastically and the company started dealing in war surplus. Electronic parts and assemblies were advertised in a monthly flier which described the items and proposed applications. The first electronic kit was the 0-1 oscilloscope, which was developed to use surplus CRTs. After additional test equipment kits, audio gear was

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introduced in 1952. The line had increased to 60 kits when amateur radio gear was introduced in 1954, the same year Howard Anthony was killed in an aircraft crash.

The company was sold to Daystrom, Inc. in 1955, and continued to grow. In 1962, Schlumberger purchased Heath. The catalog expanded to offer more than 300 kits, including the H8 and H11 computers in 1977. Zenith bought Heath in 1979 for the computer business, and moved kit sourcing and packing to Hong Kong. In 1989, Heath was sold to the French firm Groupe Bull, which decided to concentrate on education, home security and home automation.



The brief history, while far from complete, provides an overview of key Heath people, facilities and corporate relationships. One of the things that struck me was that, even though kit builders were involved with Heath products at the personal level, Heath grew to be a very large business. Once it attained reasonable size, the company had to operate under the same constraints as any other large organization. That Heath's personal involvement with customers could be sustained for so many years is a tribute to the many people involved in designing, developing, marketing and supporting Heath products. Terry Perdue doesn't chronicle all of the kits and products offered by Heath throughout its history; perhaps someone else will provide a detailed history of Heath's wares.

Most of the pictures in *Heath Nostalgia* are of people or Heath facilities, and a few early advertisements are included. The graphics are reproduced with a large grain, which, while not too obtrusive in the photographs, renders the smaller print in the advertisements illegible. If poster-size reproductions of some of the early literature were made available, they would make interesting collectibles.

Continued on page 71



# **SB** Mailbox

# SOUNDS FISHY

My home system consists of a pair of Aria 5s and an Eminence 15" dual-voice-coil subwoofer in a sealed box. I use separate amplifiers and a 24dB/octave electronic crossover set to 100Hz. Each low-frequency channel powers its own voice coil in the subwoofer.

Originally, 1 used the subwoofer in the front-firing position, but found the subjective room response lacking in the low bass and excessive in the midbass. As I thought about possible remedies, 1 recalled that the closemiked frequency response of the Audio Concepts Sub 1 was essentially a fairly narrow hump centered on 50Hz, but the room response was extremely flat when the unit was used as designed with the driver very close to the floor.

I positioned my subwoofer, driver-down, in exactly the same place as before. I propped it up on three tuna fish cans to give the driver frame about 2" of clearance from the wood floor. The subjective room response improved phenomenally. I was able to turn down the gain of the subwoofer's amplifier and enjoy smooth, extended response without the 80Hz hump that had resulted previously from the subwoofer's rising room response.

What is this effect whereby the low bass of a subwoofer is boosted by very close placement to the floor?

Kurt Rosenfeld New York, NY 10003

# **SMOKING BAN**

Randy Parker's article in *SB* 4/93 on "The Prism V Satellite/JBL Subwoofer" (Part 1, p. 16) was excellent. Technically, it has something for everyone and covers a wide range of skills. The author, with the tech support of David Clark, has done a great job with the passive and active crossover designs for this loudspeaker system.

The passive crossover design uses a 30µF capacitor in the parallel resonant circuit L4-R2-C5, providing a notch filter. These circuits are widely used to make small corrections due to undesirable acoustic responses.

Above this notch filter's resonance frequency, the capacitor carries all of the load current (i.e., the audio frequency current into





the driver). This current can be quite substantial at high power levels: in the order of 1–3A for a loudly played brass choir passage into a  $4\Omega$  midrange driver (not necessarily the case in this article).

Capacitors used in impedance-correcting networks for woofers and midrange drivers may also be subject to high audio currents. Those used for tweeter crossover sections have dramatically less current flowing through them, because the energy in that portion of the speech and music spectrum is quite small, and their generally lower capacitance values open up a wider selection of device types.

The big caps, such as the 30µF device used in the article's passive notch filter, are nearly always nonpolarized electrolytics. Audio current of 1-3A through one of these demands selection of a low-ESR (equivalent series resistance) type. Audio frequency current is called "ripple current," and each capacitor design-the manufacturer's part-number series-nearly always has a maximum continuous amount specified in the data sheet. Exceeding the manufacturer's limit repeatedly for long periods will severely damage or destroy the capacitor. Some readers will remember the bad press the early switch-mode power supplies received 20 years ago due to excessive ripple current failures.

Many SB readers buy components from Digi-Key in Minneapolis. Glancing at the capacitor section in their catalog is revealing.





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From their Catalog #935, I found that the Panasonic NHE (an axial-lead aluminum electrolytic) is listed as value/volts = 33/25, with maximum ripple current of 1.3A. Although the spec is included for the guidance of switch-mode power supply designers at 100kHz, the IkHz value is nearly the same.

In a recent loudspeaker design, I needed to use two paralleled sets of back-to-back Panasonic NHEs to provide for the correcting network's longevity with a clear conscience. This speaker system is for high-powered speech reinforcement and will be in a very difficult location to maintain. I don't want to smoke a cap in that application, and I imagine you don't want to smoke one at home, either!

Dick Campbell WPI/ECE Acoustics Lab Worcester, MA 01609

# **DIRECTION FINDER?**

I have been looking in past and present issues of *SB* for information dealing with the construction of in-wall speakers. I know this is a less than ideal situation, but since I am doing new construction I thought I might have more control. In any event, could you point me in the right direction with articles or sources for plans? I have heard of using "Great Stuff" foam to create sealed volume between studs/joints, although controlling the exact internal volume must be difficult. Thank you for any assistance you can provide.

Steve Treat Yarmouth, ME 04096

# FOR GOOD MEASURE

When describing the RATE function in Part II of his 1MP series (*SB* 2/93, p. 30), Bill Waslo states that the higher sampling rate of 61.44kHz can be used for measuring frequencies above 650Hz. Is this low-frequency limit due to generalized room boundary conditions, or is it a function of the acquisition/FFT routines?

# PREVIEW Audio Amateur

### Issue 3, 1993

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WRITE OR CALL FOR A FREE CATALOG Reader Service #9 I'm interested in making off-axis measurements. Intuitively, this does not seem possible using near-field techniques. Of course, I'll have to go outdoors, but this eliminates roomboundary conditions. So it would be nice to make full-range measurements at a single sampling rate.

In addition, I would like to make full-range impedance measurements so I can import the impedance/phase curves into LEAP, which also has a T/S parameter calculation routine. (LEAP uses a frequency dependent motor model to describe a driver's high-frequency impedance rise.) If I can indeed make full-range acoustic measurements with the IMP set to the higher sampling rate, can I also make full-range electrical/impedance measurements?

Bill Waslo deserves a lot of praise for introducing an affordable measurement device, as well as his series of articles, to amateur speaker builders! Many thanks.

Don Vogel APO AE 09180

Bill Waslo responds:

Actually, the implied low-frequency limit is due to imprecise writing on my part. What I meant to get across was that the lower sample rate of 1.92kHz could not be used for measuring frequencies above 650Hz. Above that, the higher sample rate must be used; however, this rate can be used to measure down to 30Hz or so, provided the largest sample SIZE is used and the data is not truncated to remove echoes.

The upper and lower limits of the frequency range over which measurements can be made are determined by the sample rate and the sampling duration, respectively. Onehalf of the sample rate, known as the Nyquist frequency, represents an absolute limit to the highest measurable baseband frequency. Assuming a perfect brick-wall antialiasing filter was wired ahead of an analog-to-digital (A/D) sampling at 61.4kHz, the highest frequency you would have any chance of measuring would be 30.7kHz. Less than ideal real-world filters reduce this upper limit to something less than half the sampling frequency. IMP reports results only up to approximately one-third the sample rate, or about 21kHz for the higher sample rate.

IMP can collect up to 4,095 samples. At the upper sample rate of 61.4kHz, each sample represents 1/61,400 seconds; 4,095 of them amounts to a duration of 4,095/61,400 = 0.0667seconds. The lowest frequency about which any data could possibly be valid will then be 1/0.0667 = 15Hz (one cycle of 15Hz energy is 66.7ms long). The measurement resolution will



Reader Service #23

also be 1511z (i.e., you will get data output at frequency points spaced 1511z apart).

It would be best not to trust the first point; after all, IMP only got to look at a single cycle of it. So the lower frequency limit for the higher rate is 3011z. Similar calculations can be performed using the lower sample rate of 1.92kHz, yielding data results below 1Hz and resolution below 1/2Hz. Due to capacitive coupling in the IMP analog sections, the lowest frequency reported is 211z. These results hold for acoustic, electrical, or impedance measurements.

In most cases, IMP will attempt to suppress plotting data about which it can have no knowledge. You may have noticed that frequency response and waterfall plots generally have no data plotted below certain frequencies. IMP performs the FFT only on real data between the time markers, so the time duration between those markers represents a hard limit on the lower frequency about which the derived frequency response can have any meaning. IMP reports no data below this frequency. On the waterfall plots, the time span which is transformed is shortened on each successive trace; hence the left edge of the trace will often advance forward as the waterfall plot progresses.

Concerning the making of off-axis measurements, I agree that near-field techniques do not seem appropriate, but you shouldn't have to go outdoors. You can make the offaxis measurements just like the on-axis ones, removing the echoes digitally. I do this by simply rotating the speaker and keeping the microphone fixed. Off-axis measurements at low frequencies will not likely yield much in the way of results. The bass is pretty much

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Regarding full-range impedance measurements: you can make them, but the 15Hz resolution obtainable with the higher sample rate won't give LEAP's routines much to work with. A better solution might be to combine data from two impedance measurements of the same device, one run at a low sample rate (for resolution at the low end) and one at the higher rate. This could be done via a spreadsheet program, or perhaps a simple utility could be written in Basic to perform the function. I'm not familiar with the format LEAP requires, so you will have to check your documentation to see whether or not this is practical.

Thanks for your kind words regarding my IMP project. I'm glad you find it useful, and I hope it leads to some novel designs.

# INCONTESTABLE FACTOR

I have some comments to offer on the Galo/Crawford exchange concerning damping factor, as it appeared in *SB* 6/92 ("Damping Factor Dialog," p. 46). I believe that speculations on the affect of damping factor on driver behavior are a holdover from vacuum tube days, before Thiele/Small developed a rational approach to the design of loudspeaker systems' bass performance. For example, if a woofer system is designed following T/S procedures, the effects of various resistances (amplifier source, interconnection cable, and passive crossover components) will all have been recognized in the calculation of the electrical Q.

In "Thiele, Small, and Vented Loudspeaker Design" (*SB* 4/80, p. 7), Robert Bullock describes exactly how to arrive at an effective electrical Q. He also quotes Thiele as stating that response variations will not exceed 0.4dB if the driver electrical Q is used without correction, provided the sum of the amplifier source resistance and other series resistive elements are less than 5% of the nominal driver impedance.

In Mr. Galo's example, he used 16 AWG wire to connect the speaker to the amplifier and was shocked to note that the effective damping factor dropped from 400 to 56. Thiele's rule, however, would permit the sum of the amplifier source and interconnection resistances to be  $0.4\Omega$ , which would result in an effective damping factor as low as 20!

Do not interpret from the above discussion that I think amplifiers with low damping factors are OK. On the contrary, it is obvious they should be high, in order that the effective



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World Radio History

damping remains high even after all of the intervening resistances have been added.

An amplifier should have a high damping factor (i.e., low internal resistance) for yet another reason: it must drive speaker systems whose impedance varies widely as a function of frequency—often over a five-to-one range. While doing so, it must maintain a flat frequency response. To do this, it must have a constant voltage source, which implies that its internal resistance is very low. For the unfortunate affect on frequency response of a low-damping-factor amplifier driving a speaker, see the January 1991 issue of *Stereophile* (p. 231).

David J. Meraner Scotia, NY 12302

#### Satellites

continued from page 28

hurting the high end (*Fig. 4*). I also tried a zobel to flatten the woofers' response. The schematic does not contain a misprint—l accidentally wired up the zobel incorrectly and found that it improved the response. Don't ask me why, but it works.

A simple and cheap way to hang the units is with angled carpet edging. With one piece

on the cabinet and one on the wall, it is very secure. By also routing a recess in the back of the cabinet, you can achieve a very tight fit to the wall (*Photo 3*).

These units sound great. I mated them with two 2 ft.<sup>3</sup>, 12" subwoofers, crossed over at 180Hz. As I had hoped, the tweeter's heightened response overcomes the added midbass caused by the corner placement. Even at lowlevel listening, the treble is crisp and clear. With the system hooked up to the TV, voice reproduction is very natural (*Photo 4*). In fact, it sounds so good that I never watch TV without running it through the stereo. MTS stereo and Dolby surround have turned me into a video freak—but that's the subject of another article.

#### **Book Report**

continued from page 64

The largest section of the book, "Memory Miscellany," is a collection of remembrances from over a dozen contributors spanning the period from Howard Anthony's teenage foray into the radio business to the HERO 1 robot's exploits passing through airport security and other experiences of the 1980s. The early aircraft business is examined along with the electronic kits which built Heath's reputation among my generation. "Miscellany" covers a wide territory, including product development, business practices, fabrication, marketing and speculation on Heath's future. Some contributions are serious, others humorous. Heath seems to have had more than its share of characters and practical jokers.

While touching on most of Heath's product lines, the recollections seem to emphasize test equipment and radio amateur gear. Perhaps this reflects the author's amateur radio background. I'd have liked greater coverage of audio equipment.

Before receiving the review copy, I expected a history book. After realizing that reality didn't match expectations, I sat back and enjoyed *Heath Nostalgia* for what it is. The title was aptly chosen.

The golden age of electronic kit building is gone. In the past, Heath shifted its product emphasis to meet the challenge of change. Let's hope present management has the wisdom to provide and support useful products desired by customers. In the meantime, we'll have our memories and a collection of stillfunctioning gear we assembled ourselves.

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+ + +

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THE INLAND EMPIRE AUDIO SOCIETY (soon to become) THE SOUTHERN CALIFORNIA AUDIO SOCIETY—SCAS is now inviting audiophiles from all areas of Southern California and abroad to join our serious pursuit for that elusive sonic truth through our meetings and the IEAS' official speaker, *The Reference* Newsletter. For information write or call, Frank Manrique, President, 1219 Fulbright Ave., Redlands, CA 92373. (714) 793-9209.

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ESL BUILDERS GROUP is a new address for people who have built or want to build ELECTRO-STATIC LOUDSPEAKERS and ASSOCIATED (TUBE) DRIVERS, or are just interested. We will concentrate on ESL-related building projects but also look at the theoretical aspects of acoustics and electronics. Interested? An answer is ensured, if you include some kind of compensation for postage and handling. Write to: Gunter Roehricht, Buhler STR.21, 7030 Boblingen, Germany.



THE WESTERN NEW YORK Audio Society is an active, long established club located in the Buffalo area. We issue a newsletter and hold meetings the first Tuesday of every month. Our meetings attract many prominent manufacturers of audio related equipment. We are involved in all facets of audio—from building/modifying to exposure to the newest high-end gear, and the chance to hear more types of music. For information regarding our society, please write to WNY Audio Society, PO Box 312, N. Tonawanda, NY 14120.

SOUTHEASTERN MICHIGAN WOOFER AND TWEETER MARCHING SOCIETY (SMWTMS). Detroit area audio construction club. Meetings every two months featuring serious lectures, design analyses, digital audio, A-B listening tests, equipment clinics, recording studio visits, and audio fun. The club journal is *LC*, *The SMWTMS Network*. Corresponding member's subscription available. Call (313) 544-8453 or write David Carlstrom, SMWTMS, PO Box 721464, Berkley, MI 48072-0464.

THE ATLANTA AUDIO SOCIETY is dedicated to furnishing pleasure and education for people with a common interest in fine music and audio equipment. Monthly meetings often feature guest speakers from the audio manufacturing and recording industry. Members receive a monthly newsletter. Call: Chuck Bruce, (404) 876-5659, or Eddie Carter, (404) 847-9296, or write: A.A.S., 4266 Roswell Rd. N.E., K-4, Atlanta, GA 30342-3738.

# ™114-S

Neodymium Magnet DPC Cone 4" Woofer

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		ST.		

Specification	
Overall Dimensions	Ø118mm (4.64") x 58mm(2.29")
Mounting Baffle Hole Diameter	Ø95mm (3.75")
	pe, Vented, Neodymium Magnet
Nominal Power Handling (Din)	150W
Transient Power - 10ms	800W
Voice Coil Diameter	54mm (2.125")
Voice Coil Type/Former	Hexatech Aluminium
Frequency Response	55-7000 Hz
FS - Resonant Frequency	65 Hz
Sensitivity 1W/1m	87 dB
Z - Nominal Impedance	8 ohins
RE - DC Resistance	5.6 ohms
LBM - Voice Coil Inductance @ 1	
Magnetic Gap Width	1.25mm (0.050")
HE - Magnetic Gap Height	6mm (0.236")
Voice Coil Height	12mm (0.472")
X - Max. Linear Excursion	3mm
B - Flux Density	0.88T
BL Product (BXL)	6.75
Oms - Mechanical O Factor	2.32
Qes - Electrical Q Factor	0.36
Q/T - Total Q Factor	0.31
Vas - Equivalent Cas Air Load	3.18 litres (0.113 cu. ft.)
MMS - Moving Mass	7.00gm
CMS	807µm/n
SD - Effective Cone/Dome Area	53cm² (20.86 sq. in.)
	PC (Damped Polymer Composite)
Nett Weight	0.500 kg

Specifications given are as after at least 45 minutes of high power, low frequency running, or 24 hours normal power operation.

The 114-S is the first of Morel's new generation of woofers, featuring a powerful Neodymium magnet system which provides increased sensitivity, lower Qt and reduced distortion. For a 4" driver it is unique in having a large 54mm (2.125") diameter Hexatech aluminium voice coil.

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THE COLORADO AUDIO SOCIETY is a group of audio enthusiasts dedicated to the pursuit of music and audiophile arts in the Rocky Mountain region. We offer a comprehensive annual journal, five bi-monthly newsletters, plus participation in meetings and lectures. For more information, send SASE to: CAS, 11685 W 22nd St., Lakewood, CO 80215, (303) 231-9978.

TUBE AUDIO ENTHUSIASTS. Northern California club meets every other month. For next meeting announcement send a self-addressed, stamped #10 envelope to Tim Eding, PO Box 611662, San Jose, CA 95161.



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THE BOSTON AUDIO SOCIETY the nation's oldest (founded 1972), seeks new members. Dues includes the monthly meeting notice and our newsletter, the BAS Speaker (6 times/year). Recent issues cover Carver, a/d/s; the founder of Tech Hi-Fi; Photo CD; plus visits from famous speaker designers; listening tests; measurement clinics; research investigations; and more. Back volumes available. Membership includes engineers, journalists, consultants, and music-loving audiophiles like yourself. For information write to PO Box 211, Boston, MA 02126-0002, USA.



THE PRAIRIE STATE AUDIO CONSTRUC-TION SOCIETY. (PSACS) meets every other month. Meetings feature audio construction, design, and analyses, blind listening tests, equipment clinics, autosound, lectures from manufacturers and reviewers. PSACS, PO Box 482, Cary, IL 60013, call Tom, (708) 248-3377 days, (708) 516-0170 eves.

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