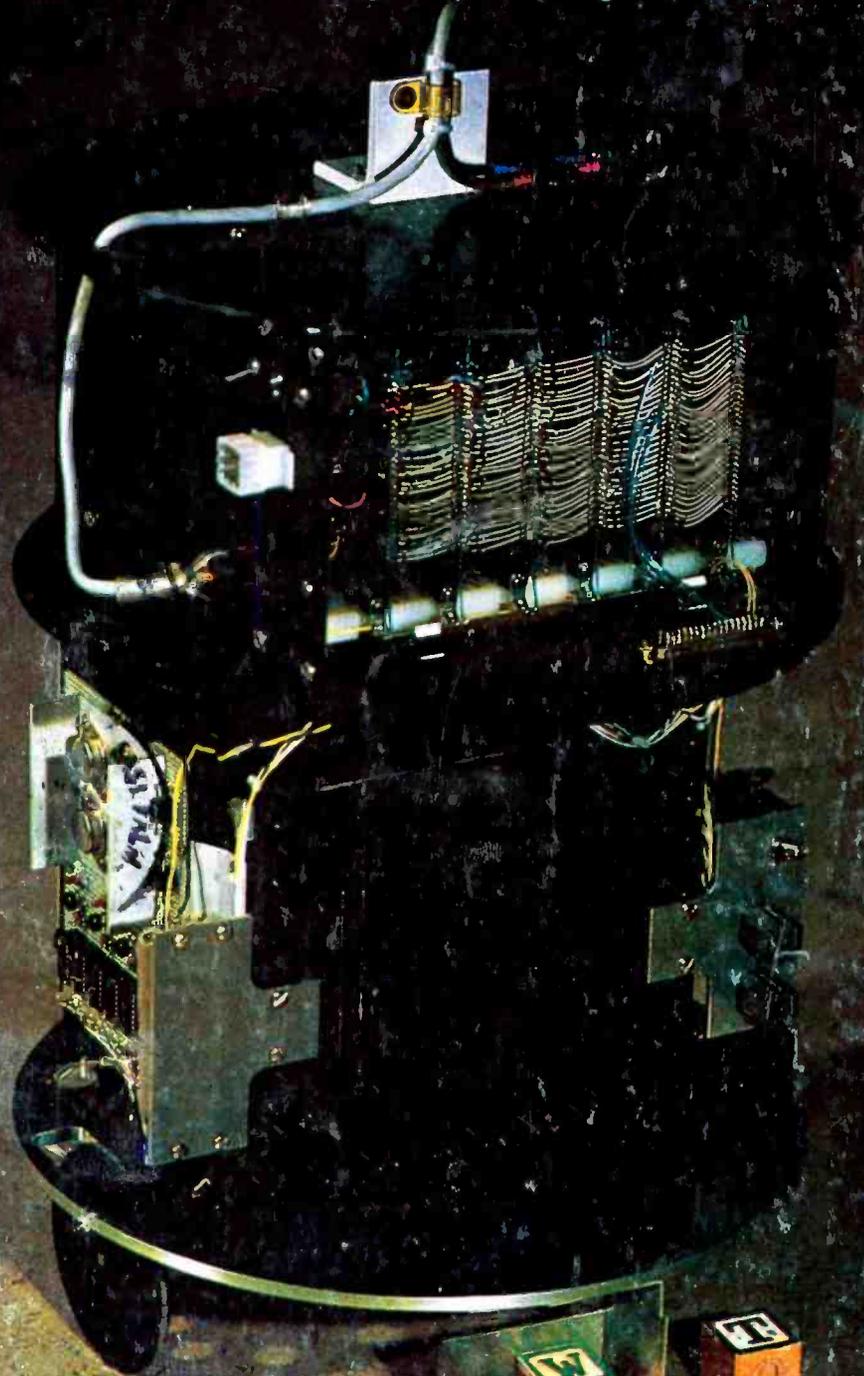


BYTE

the small systems journal

**A Mobile,
Cognitive Robot**



CT-64 TERMINAL SYSTEM



- * 64 OR 32 CHARACTERS PER LINE
- * UPPER AND lower case LETTERS
- * FULL 8 BIT MEMORY
- * 128 CHARACTER ASCII SET
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- * PRINTS CONTROL CHARACTERS
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COMPLETE WITH — Chassis and cover, cursor control, 110-1200 Baud serial interface and keyboard. Optional monitor show in photo available.

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CT-64 Terminal Kit	\$325.00
MM-1 Monitor (assembled)	\$175.00



219 W. Rhapsody

San Antonio, Texas 78216

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- Send the MM-1 monitor too. Send Data
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You'll be impressed with the Z-2's low price, technical excellence and quality. So see it right away at your computer store—or order directly from the factory.

Z-2 COMPUTER SYSTEM KIT (MODEL Z-2K) (includes 4 MHz μ P card, full-length 21-card-slot motherboard, power supply, one card socket and card-guide set, and front panel; for rack mounting) \$595.

Z-2 COMPUTER SYSTEM ASSEMBLED (MODEL Z-2W) (includes the above as well as all 21 sockets and card guides and a cooling fan; for rack mounting) . . . \$995.



Shown with optional bench cabinet

*kit price



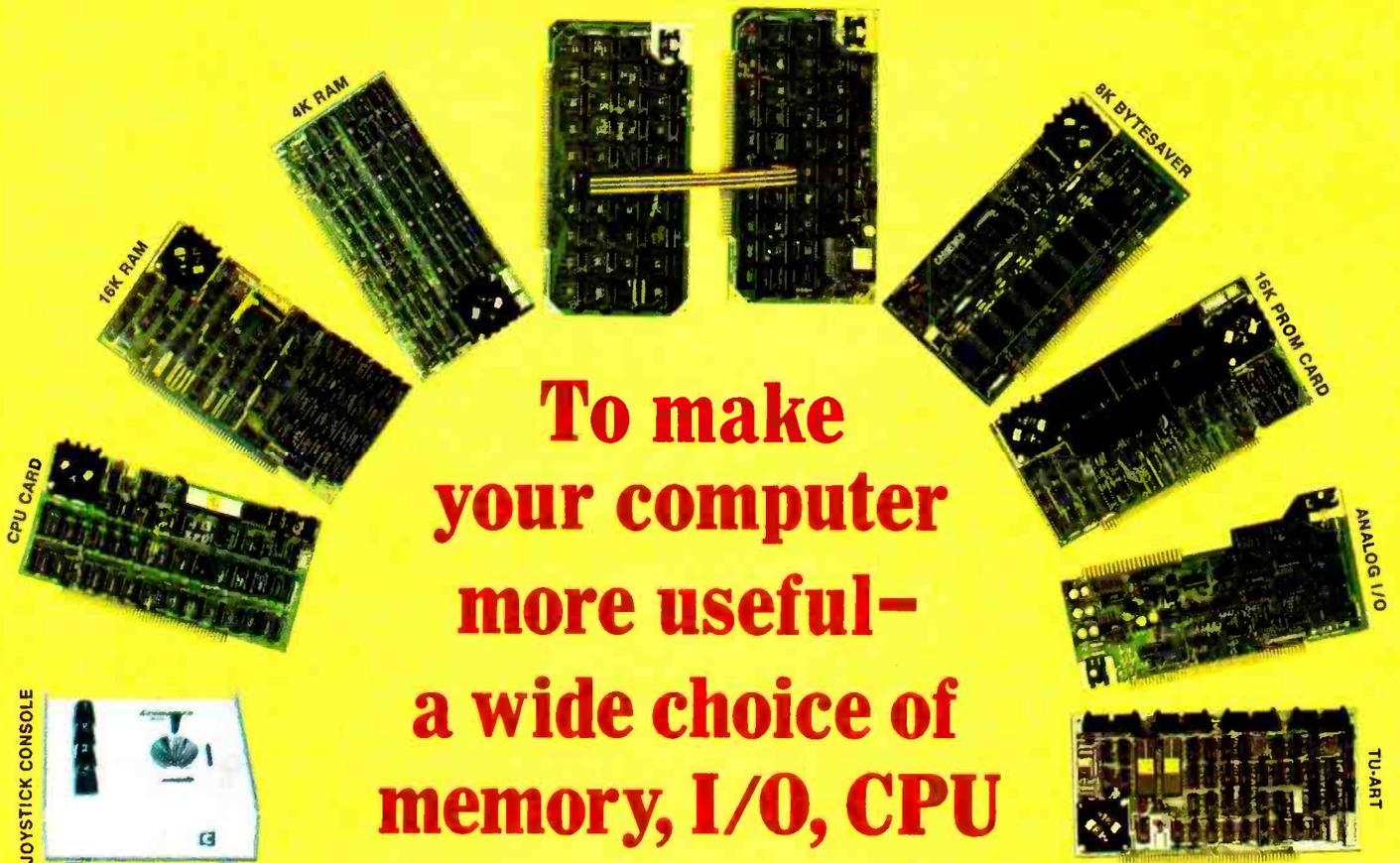
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In This **BYTE**

Some uses of a microprocessor involve the connection to the outside world through an analog interface. When fooling around with such projects from music generation to robotic control, however, it quickly becomes necessary to have a large number of inexpensive real world interfaces. To help point you in the right directions Douglas R Kraul supplied an article on **Designing Multichannel Analog Interfaces**.

In the past, readers have seen some interest expressed in the concepts of robotics, the use of small computers as the brains of mobile automated mechanisms. Robots have long been fancied in science fiction literature and cinema, but only rarely have people taken any practical steps towards a "real" robot as opposed to paper romanticisms or stage dummies. One of those rare cases is that provided by Ralph Hollis and his associate Dennis Toms, both of whom are physicists at the University of Colorado, Duane Physical Laboratory, Boulder CO. Ralph has been pursuing the design of practical robots as an avocation since 1957, and lately has progressed to the point of a working mobile computer system called Newt, whose picture provides the theme of this month's cover. Turn to Ralph's article, **Newt: A Mobile, Cognitive Robot** for essential background information on contemporary robot design philosophies.

Hard copy is a most useful output, but it tends to be somewhat expensive. Dan Fylstra shows one very attractive option in his article on **Interfacing the IBM Selectric Keyboard Printer**. Dan purchased a used print mechanism late in 1976, and since then has successfully interfaced the device to his KIM-1 system. Readers interested in using these printers (which are available in significant numbers on surplus markets) will find Dan's article an essential guide to the art.

How can hardware be used to accomplish the details of **Interfacing With an Analog World**? Turn to author Joseph Carr's second part of a two part series to find out some of the details of basic conversion circuits which use the outputs of sensors and preamplifiers discussed in last month's article.

Much of the software that is available on the market today is available on paper tape so as to be easily read into your microprocessor. The problem is that most common paper tape readers are so slow that it seems to take forever to read a large program into memory. In the article **Come Fly With KIM**, Rick Simpson introduces us to a solution to this speed problem: the Fly Reader, which he uses with MOS Technology's KIM-1.

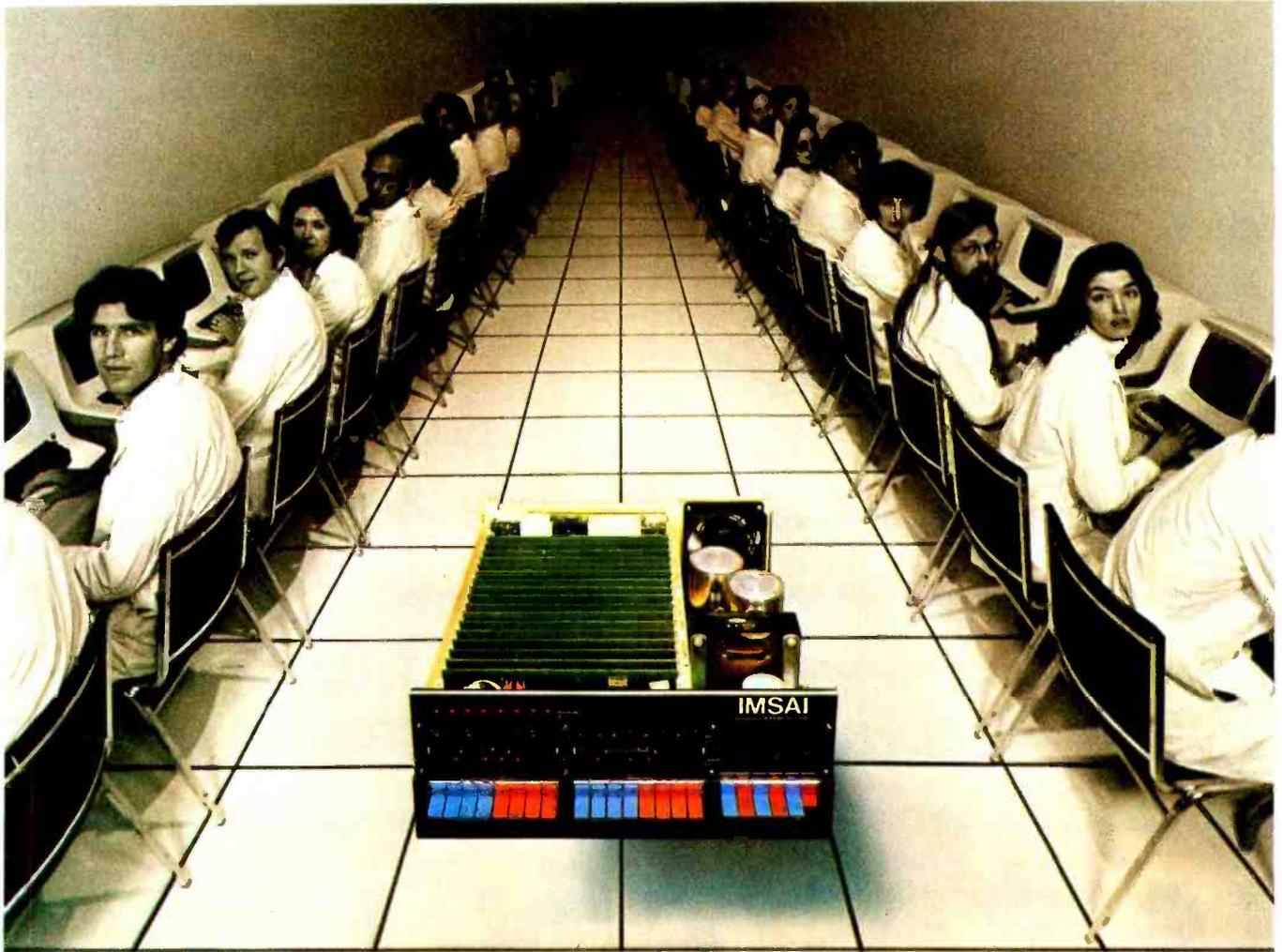
Now that you've got the hardware built, how do you run it? Ken Welles answers this question in **Software for the Economy Floppy Disk**. His previous article (February 1977 **BYTE**, page 34) described how to construct an inexpensive floppy disk with minimal hardware. This month he provides a series of subroutines to run it, which could easily be expanded into a complete floppy disk operating system.

Last month in the first part of his article **Artificial Intelligence, An Evolutionary Idea**, Michael Wimble introduced us to the use of a simulated evolution technique by which it was possible for a program to alter itself and reshape its responses as a direct result of an outside stimulus. This month in **Part 2: Implementation**, Mr Wimble details how the computer experimenter can implement this type of program on any small computer system.

To many people the concept of assembly language is that of the fundamental language of the computer next to machine language. However, each particular assembly language command must be broken down into a series of simpler command sequences. These commands are known as microinstructions. In his article, **An Introduction to Microprogramming**, S M Quek describes how the concept of microinstructions is a great benefit to the user of a computer, allowing the easy change of basic instructions.

In previous issues Michael McNatt has shown us the availability of Baudot teleprinters and the ways in which they can be interfaced with your microprocessor. In his concluding article, **A Guide to Baudot Machines: Part 3, A Teleprinter Test Circuit**, he describes a test circuit that can be used for generating Baudot characters for alignment and adjustment purposes.

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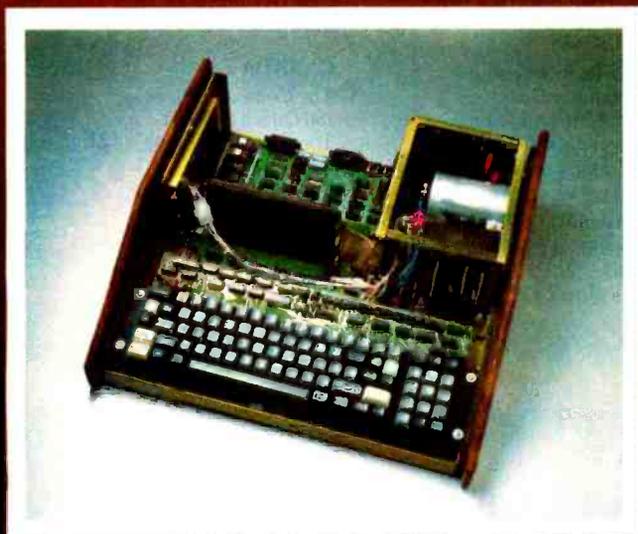
Sol System II has the same equipment plus a larger



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The Software Dilemma:

Editorial

How is it possible to simultaneously make software widely available (and low priced), yet reward the producers of good software with adequate compensation for their efforts?

By Carl Helmers

Conventional wisdom has it that proprietary software must come at extremely high prices, commensurate with concentrated work on the part of a small number of dedicated and thoughtful programmers. After all, this wisdom has it, we'll only sell a few copies of package X anyway, so why not keep a tight lid on it and charge as much as possible?

This conventional wisdom has worked well in the past, when the typical computer system might cost upwards of \$10,000 or \$100,000. But when the typical computer system comes in at a price on the order of \$1000, paying prices which are of this same order of magnitude for software packages is not a very likely move on the part of the individual purchaser with his or her personal budget.

In the personal computing field we are participating in a market phenomenon characterized by a change from the situation which supports the conventional software wisdom, to a new situation which has its own characteristics. More and more people are getting into the swing of things with computer use, and thus more and more people have needs which can and should be filled by specialized software products. Where computers are concerned, when we talk about a 100,000+ person active individual user market as we do today, we are for the first time talking about the potential for mass marketing of software in ways unheard of in the conventional wisdom of computing. Establishing a new "conventional wisdom" is clearly required; as a step toward that goal, this paper provides a survey of the prospects for mass marketing

of software, and a solution of the software dilemma posed above.

Let's Draw Some Parallels: Woodworking

Like many individuals, I dabble a bit in the arts of crafting furniture. Suppose, for example, that I want to build a nice, neat contemporary rolltop desk for my study. As an individual with limited time available for such leisure crafts activity, I'd probably want to start with an existing design rather than working out all the details myself. In seeking the end product of a rolltop desk, I'd be in the same situation (as a wood craftsman) as the owner of a computer system desiring a compiler, assembler, application product or peripheral. I know in principle that rolltop desks exist and that in principle I could design then fabricate one, or use an existing one as a mental model with my own variations. But to save time and possible mistakes I might want to find some source of a "proven" design with detailed information on achieving the goal of a rolltop desk. Well, in the world of woodworking, as in the world of photography, the world of live steam model engines, or the world of backpacking, there are numerous sources of information including ready-made designs and techniques. I refer, of course, to books which are just published products with specific orientation or theme.

Similarly, when I have a computer system and I know that some neat language or software development tool exists, I also know that in principle I could write such a package myself using my own design or general design concepts taken from any

This editorial consists of the text of a paper delivered at the First West Coast Computer Faire in April of this year.

About those missing mailing wrappers and the May issue:

A strike at the printing plant was responsible for May BYTEs arriving late to subscribers and for May and June issues being mailed without the customary brown wrappers. The wrappers will be restored as soon as our printing situation is restored to normalcy. ■

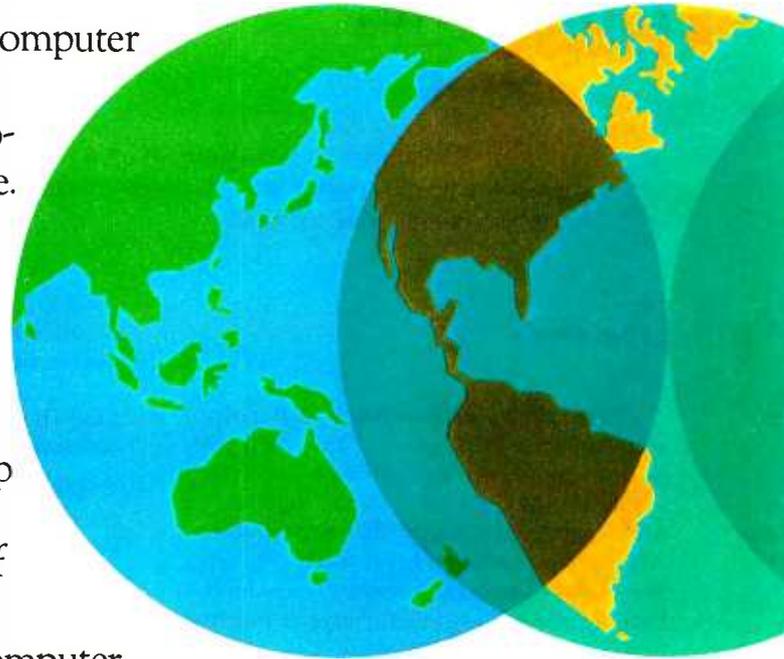
Continued on page 68

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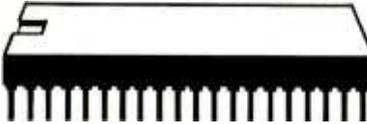
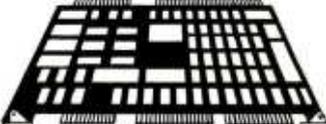
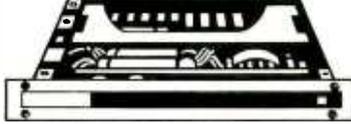
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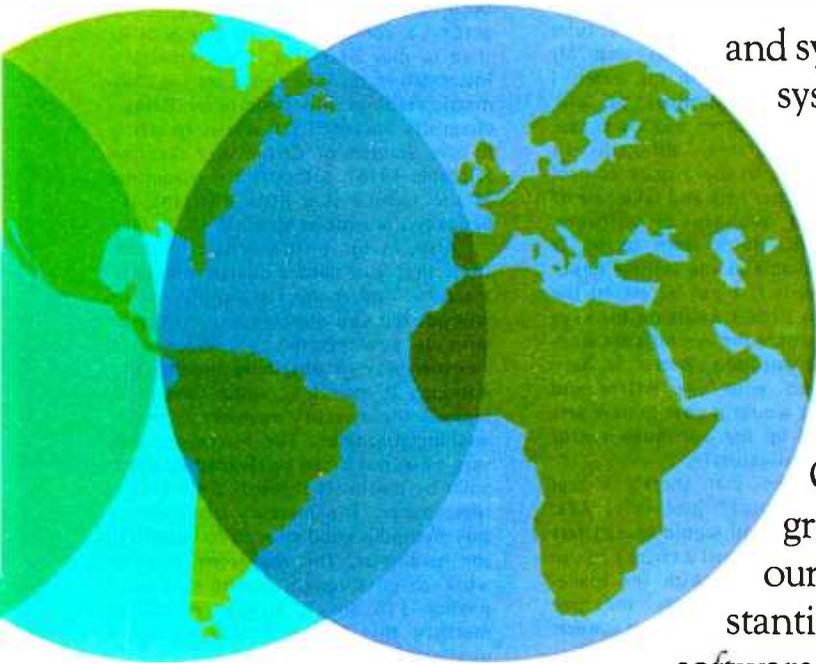


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Letters

AN APL LOVER'S STORY

Regarding the two letters by APL enthusiasts in your February 1977 issue, let me lend my voice to this group. While still in high school, I got my first taste of APL from a friend at IBM via a long-distance line to Detroit, and got access to Xerox Sigma 9 APL while a junior in college. Since then I've written a lot of APL code — including a 7 page pattern recognizer program (imagine what that would be in BASIC!) and about five pages of n-dimensional optimization routines, and now I am running IBM 5100 APL where I work. Needless to say, I think APL is the greatest thing since left-hand Turing machines.

I have just bought an ECD Micro-Mind computer (graphics) system with the explicit intent of buying 32 K of memory when the price comes down, and beg, borrow, buy, or if necessary, write an APL interpreter. (What about Tiny APL, analogous to Tiny BASICs being written now?) Since my computer will have graphics capability, I won't be interested in APL ROMs. (Might I suggest mnemonics, eg: \$R for APL "rho," \$QQ for APL "character quad.") I have used them "without hardly noticing." But in any case, put my vote in for APL, and I would be happy to hear from any APL enthusiasts.

Gregg Williams
3439 Southern, #7
Memphis TN 38111

SOME APL PERIPHERALS QUESTIONS

You can add my name to the list of those who would be interested in an APL character generator chip (Letters, February 1977). Like a lot of people who have used APL, I caught the bug, and have been disappointed that there is no software for the 8080 to support APL. Though I suspect it's only a matter of time.

A cost of \$20 to \$25 for a chip "feels" right to me. This would require about half the 500 buyers that Mr Montgomery postulates in his letter. I have a couple of warnings to add, though. First, a full upper and lower case keyboard is desirable, although it's certainly possible to use a spare control key to signal upper case and take care of the translation problem in software. Such keyboards don't seem to be as cheap and available as the surplus upper case ones. Second, I can attest to the fact that using little stickers on the keys to show you where the symbols are leads to a lot of frustration. Better to have keys imprinted with the letters and symbols. What would it cost to have sets of these made up for distribution with the character generators?

I'm convinced that there's a real market for a "small" (not tiny) APL interpreter. The word would spread fast to those who haven't had a chance to use it and are putting up with the inadequacies of BASIC (mainly the size of source programs) without much complaint.

James C Wilson
Ketron Inc
3250 Wing St #402
San Diego CA 92110

We know of one interpreter which is nearly complete for the TMS-9900, plus

several 8080 versions. Watch future BYTEs for some fairly extensive APL information. Articles are now in preparation concerning APL interpreter design, use of APL, etc.

ON AUTOMATED BAROMETERS AND OBTAINING MERCURY

I was very interested in Mr Firth's article on weather predictions (December 1976 BYTE, page 62), for there have been very few articles in BYTE on getting a computer to do things other than to play Star Trek. I was especially interested in his idea about getting barometric readings into a computer. Being a chemistry student I ran across an article in the *Journal of Chemical Education* (October 1976). Although the barometer in the *Journal* is a little different, it's basically the same as Mr Firth's.

Also, in Mr Firth's article he mentions that you need a quarter pound of mercury, and many lab supply houses will not sell you mercury for any reason without a company's purchase order. Another way of obtaining the mercury, although it may take some time, is to remove the mercury switches from old washing machines. The mercury is not very pure, but it can be cleaned up somewhat by passing it through a pin hole in filter paper. The mercury obtained by this method is good enough to be used in the barometer. This may seem a lot of work to go through, but it sure beats paying \$13 for a quarter pound of mercury that you will probably use a quarter of.

D Pasken
23 Farview Cir
Camillus NY 13031

Mr Pasken enclosed a Xerox shot of the article he mentions, which can be found on page 670 of the October 1976

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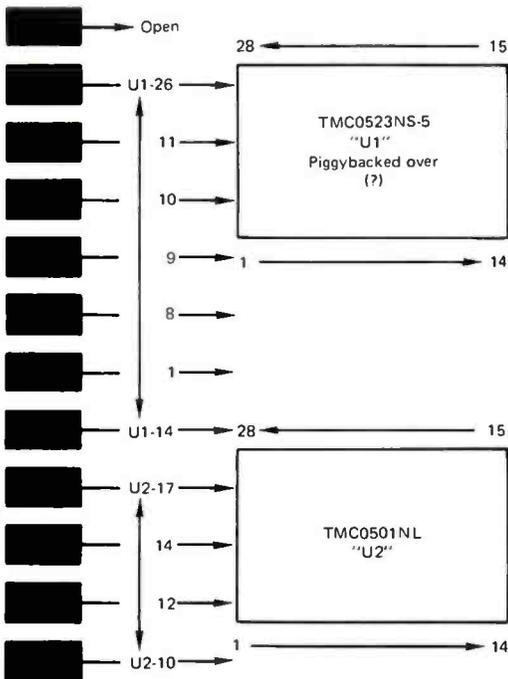
TEXAS RESIDENTS ADD 5% SALES TAX

issue of the Journal of Chemical Education. The design is by John T Viola and William E McDermott of the US Air Force Academy. The design, in detail, is a recording manometer. Reference is made to a paper, circa 1953, by H T Svec and D S Gibbs (in Rev Sci Instr, 24, 202, 1953). The paper also gives a reference to a source of wire for the measurement: 28 gauge bare nickel chrome wire cable cord manufactured by Consolidated Companies, Chicago IL.

SOME SR-51 CALCULATOR INTERFACE INFO

I read with interest Ralph Getsla's (Letters of January 1977 BYTE) request for information on interfacing his SR-52. My interests are similar, only my outlook evolves around the use of the SR-51 A terminal strip that can be seen upon removal of the battery pack of the 51. My plan here is to interface the 51 to the modified TV typewriter terminal that I am presently in the completion stages of building, I was able, so far, to track these lines back to their source by the use of a hand held flashlight after disassembly of the 51 (see interface diagram). If I can get any information on the two chips in question I believe I would be in business.

William D Lewis
469 Heatherbray Ct
San Jose CA 94136



PS: Could this interface be the same as the SR-52? I will be writing Texas Instruments for any information they will be able to give me on these devices. If not, I will operate via the probe.

Who knows?

Continued on page 122

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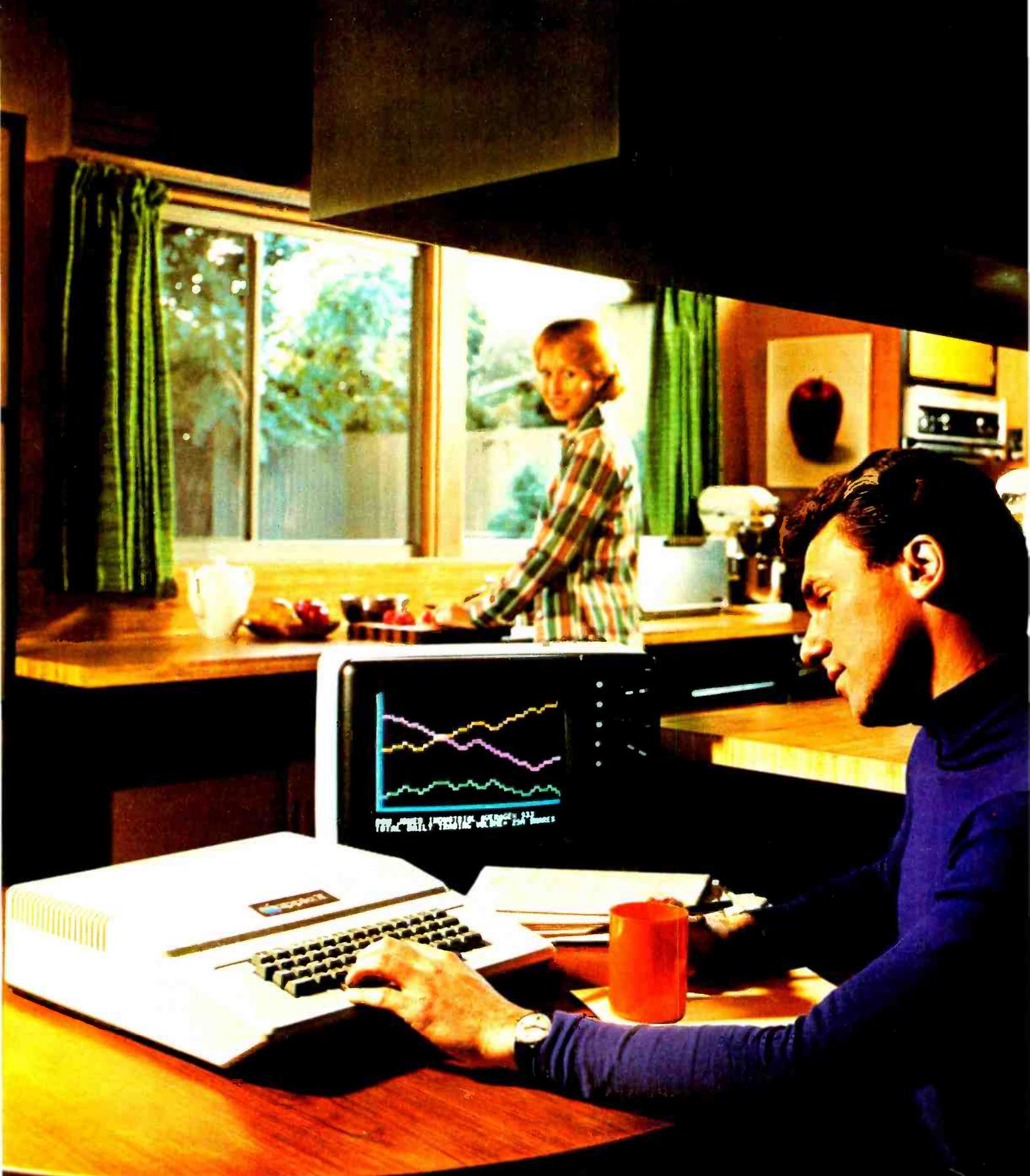
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Apple II™ is a completely self-contained computer system with BASIC in ROM, color graphics, ASCII keyboard, lightweight, efficient switching power supply and molded case. It is supplied with BASIC in ROM, up to 48K bytes of RAM, and with cassette tape, video and game I/O interfaces built-in. Also included are two game paddles and a demonstration cassette.

SPECIFICATIONS

- **Microprocessor:** 6502 (1 MHz).
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 - Text—40 characters/line, 24 lines upper case.
 - Color graphics—40h x 48v, 15 colors
 - High-resolution graphics—280h x 192v; black, white, violet, green (12K RAM minimum required)
 - Both graphics modes can be selected to include 4 lines of text at the bottom of the display area.
 - Completely transparent memory access. All color generation done digitally.
- **Memory:** up to 48K bytes on-board RAM (4K supplied)
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 - Apple game I/O connector
 - ASCII keyboard port
 - Speaker
 - Composite video output

Apple II is also available in board-only form for the do-it-yourself hobbyist. Has all of the features of the Apple II system, but does not include case, keyboard, power supply or game paddles. \$598.

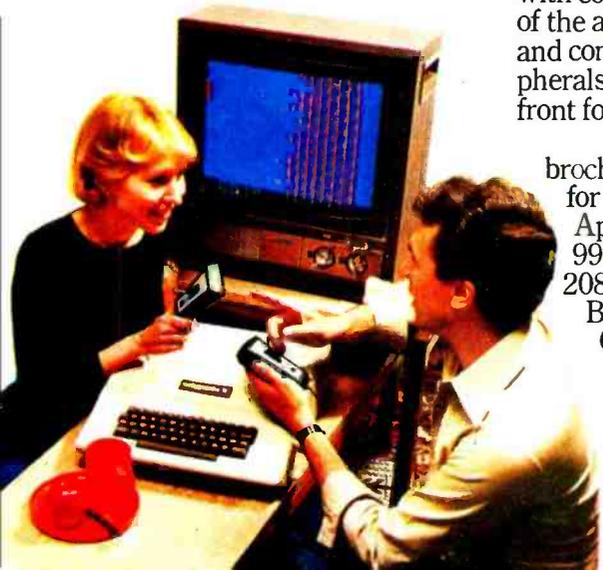
PONG is a trademark of Atari Inc.

*Apple II plugs into any standard TV using an inexpensive modulator (not supplied).



complete, ready to use computer, not a kit. At \$1298, it includes video graphics in 15 colors. It includes 8K bytes ROM and 4K bytes RAM—easily expandable to 48K bytes using 16K RAMs (see box). But you don't even need to know a RAM from a ROM to use and enjoy Apple II. For example, it's the first personal computer with a fast version of BASIC permanently stored in ROM. That means you can begin writing your own programs the first evening, even if you've had no previous computer experience.

The familiar typewriter-style keyboard makes it easy to enter your instructions. And your programs can be stored on—and retrieved from—audio cassettes, using the built-in

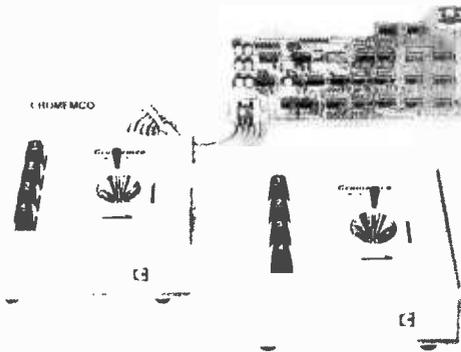


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What's New?

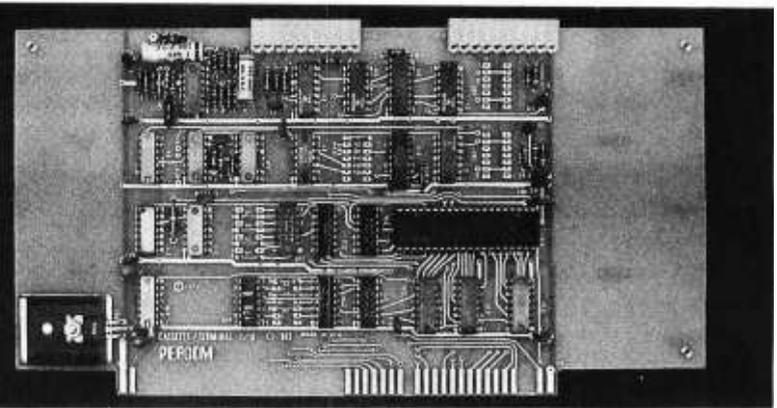
Games are More Fun with Action Inputs



This little truism can be confirmed by anyone who has implemented and played a space war with joystick control, or used joysticks for direct control of moving systems such as robots or other mechanical marvels. Cromemco, 2432 Charleston Rd, Mountain View CA

Hard Copy That's Hard to Beat for Speed

The Digital Group, POB 6528, Denver CO 80206, has announced what is probably the fastest and widest line width matrix impact printer mechanism and electronics package yet to be marketed to personal computing users. In kit form (kit refers to the electronics, not the mechanics), prices for this printer start at \$495. What you get is a fast 120 characters per second 5x7 dot matrix printer which gives 96 characters per line at 12 characters per inch pitch, and line spacing of six lines per inch.



A New Audio Tape Cassette Driver for the Altair Bus

PerCom Data Company Inc, 4021 Windsor, Garland TX 75042, has just introduced a new version of Harold Mauch's Kansas City standard phase encoding audio signal interface board, a version which plugs directly into an Altair bus slot. Harold's design allows phase encoding with redundancy at 300

bps (Kansas City standard), 600 bps, 1200 bps and 2400 bps. In addition to the tape interface function, the CI-812 product also includes a companion RS-232 terminal interface with data rates from 300 to 9600 bps. The kit price of this board is \$89.95, and an assembled version is \$119.95.■

Circle 602 on inquiry card.

94043, has sent along this photo of the new model JS-1 joystick console adapters for their Altair compatible D+7A IO board. What you get is two independent game control boxes with two axis joysticks, four game function buttons, and loudspeaker outputs for aural effects (such as photon torpedo or phaser sounds). The kit form of the box (in foreground) is available for \$65.■

Circle 601 on inquiry card.

Inexpensive Wire Wrap Tools



Since it is a true impact printer mechanism intended for computer systems use, it will take up to four part forms and should prove most useful to business people for that reason. For the amateur computer person with software development in mind, the 120 character per second rate means listings of assemblies and compilations which take one twelfth the time of a 10 character per second Teletype, but at a price for the mechanism and its interface which is lower than the new cost of a Teletype! Other features of the OEM mechanism selected by the Digital Group include built-in ribbon reiners for a total ribbon life of 10,000,000 characters, use of 8.5 inch (22 cm) wide standard roll, fan-fold or sheet paper, an 8 bit parallel interface ready to plug into your computer's output port after you wire up the cable, and the option of double width characters. For those with idle curiosity, the inking life of 10,000,000 characters before replacement of ribbon corresponds to over 23 hours of flat-out printing, or 250,000 lines with 40 non-blank characters per line.■

Circle 603 on inquiry card.

OK Machine and Tool Corporation, 3455 Conner St, Bronx NY 10475, has come out with a unique product line of wire wrapping tools and accessories for the amateur electronics person. These products include manual and battery powered wire wrapping tools, precut and stripped wire, wire rolls, dual in line package sockets, and wire wrapping "kits." Of particular interest to people on a tight budget is a new low in prices for wire wrapping tools which are powered. The OK BW-630 battery powered wire wrap tool uses standard C size batteries and comes equipped with a bit and sleeve for wrapping AWG 30 wire for only \$34.95 (less batteries). This is not a kluge, but a genuine wire wrap gun with positive indexing mechanism to return the bit to a well defined position after each wrap, and the usual "anti-backforce" spring loading of the bit to prevent overwrapping. Both of these features are standard items on the industrial wrapping guns which have been used for years.■

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Designing Multichannel

Douglas R Kraul
4373 Ashwoody Trl
Atlanta GA 30319

Analog interfaces to and from the personal computer system can present a difficult dilemma to the small systems user: The analog interface usually is a very expensive proposition, especially if more than one input and one output are needed. Schemes like that suggested by Roger Frank [page 70 of the May 1976 issue of *BYTE*], can greatly reduce hardware complexity, and thus cost, since much of the interface burden is left to the software of the computing system. Direct extension of this principle to the case of multiple input voltages and multiple output voltages can, however, result in a hardware cost that at the least rises linearly with the number of needed outputs and inputs. One alternative scheme requires an additional bit of input to the computer and one additional voltage

comparator for each additional analog to digital input up to a total of 8. On the output side each additional voltage output leads to an additional 8 bit output port and an additional 8 bit digital to analog converter. This results in a situation where a many input, many output analog interface requires an inordinate amount of hardware, which means money to the user. (We should not kid ourselves by saying that large numbers of analog channels are rarely needed. Many worthy applications, like control of analog music synthesis, automated test facilities or control of robots would easily push the number of channels needed beyond the point of no return for the previously suggested schemes of interface.) Thus one must turn to a modified philosophy of interface design in order to meet the necessary goal of a less expensive analog interface.

Fortunately digital techniques provide us with a method of solution to the problem: time multiplexing. Time multiplexing is simply the process by which one device can be made to function as many logical devices. To the user these virtual devices appear as if they were full fledged dedicated devices. Thus our objective is to find some technique by which one analog to digital converter and one digital to analog converter can be made to function in many seemingly simultaneous conversions.

Basics of Time Multiplexed Interfaces

The basic principles are illustrated first for the analog to digital case. Figure 1 illustrates the hardware that allows multiplexed analog to digital conversions. An output port, A, from the computer is used to provide the necessary eight bits to drive the digital to analog converter (DAC). The output of the digital to analog converter is connected to the minus input of the voltage

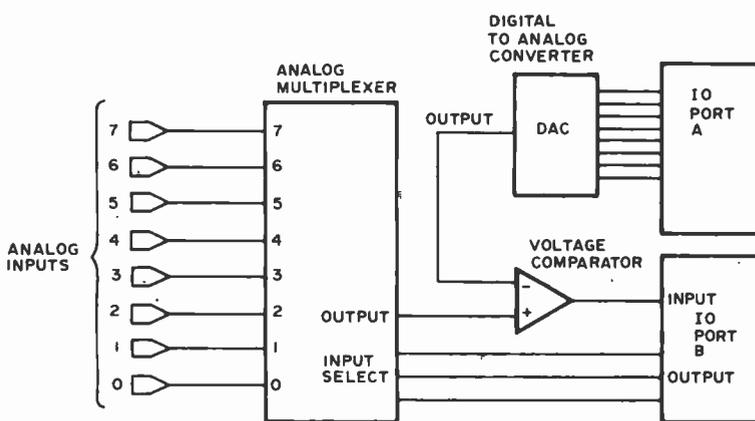


Figure 1: Block diagram symbolizing the hardware that is needed for a multiplexed analog to digital converter for eight inputs. The three outputs from IO port B select the analog channel. The output from the multiplexer is fed to the negative side of a voltage comparator. An analog output from a digital to analog converter is fed to the positive side of the comparator. When the analog value from the converter is greater than the value being tested, the voltage comparator will output a 1. Using successive approximations the input voltage can be determined.

Analog Interfaces

comparator. The output from the comparator provides an input to the computer by way of the most significant bit of an input port to the computer. This structure thus far is identical to the scheme proposed by Roger Frank. The difference is that the positive input to the comparator is no longer connected directly to the voltage to be converted. Rather the comparator is connected to the voltage to be converted by way of an analog multiplexer.

The analog multiplexer here is performing the needed function that allows one analog to digital converter to deal with many channels. A typical application might use an 8 to 1 multiplexer. Thus any one of the eight voltage inputs might become the voltage to be converted if the multiplexer selects it.

The selection is accomplished by a binary code applied to the select input of the multiplexer. In an 8 to 1 device the binary code 011 would pick the input labeled 3. The code which selects the input was set by the computer through an output port. For our 8 to 1 example three bits would be needed, possibly originating from the lower three bits of an 8 bit output port.

This change results in almost no change in the software that would service the analog to digital conversions. In fact, the only necessary modification is to preload the channel selection word, which chooses the voltage to be converted in the proper output port. Then the analog to digital conversion routine can be called.

Time Multiplexed Digital to Analog Conversion

Multiplexing of the analog to digital conversion really only solves half of the analog interface problem. The problem of economically generating multiple analog out-

puts from the converter still remains. We can apply the versatile analog multiplexer to solve this problem as well. (However, there are complications that can mask the simplicity of the method.)

The basic hardware of the multiplexed digital to analog converter output is shown in block form in figure 2. Once again the source of the digital to analog converter's word is an 8 bit output port from the computer. The output is now connected to multiplexed sample and hold circuits. Much like the multiplexer used in the analog to digital conversion system the multiplexed sample and holds connect the output of the digital to analog converter to the desired analog output which is chosen by the select inputs to the multiplexer. The difference between the plain analog multiplexer and

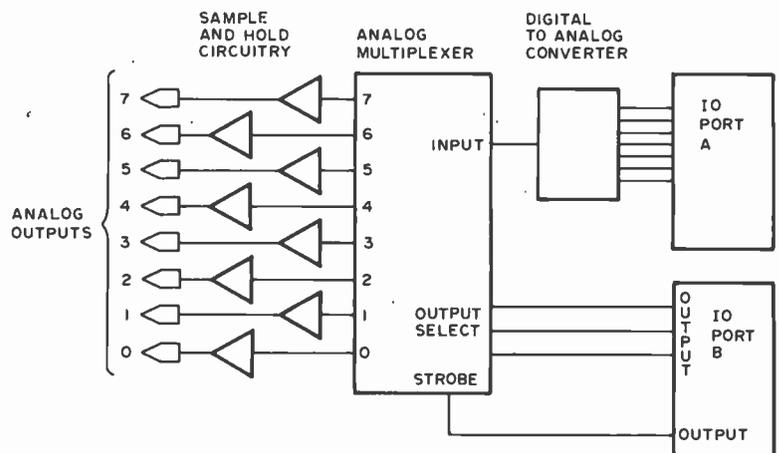


Figure 2: Block diagram of the basic hardware needed for a multiplexed digital to analog converter for eight channels of output. Three bits of output from IO port B select the channel that is to be used. When a strobe is enabled the chosen channel is activated. Each analog channel has a sample and hold circuit which must be updated periodically due to the leakage of the capacitor.

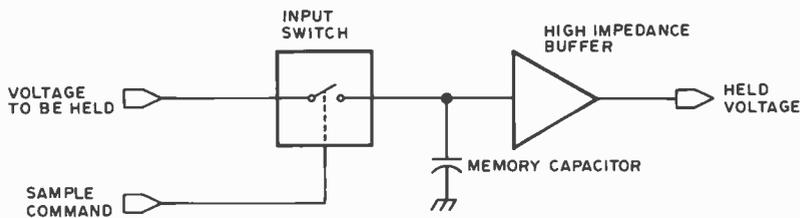


Figure 3: Functional schematic of a typical sample and hold circuit. The input switch can be mechanical or electronic. In the case of a multiplexed interface it is an analog switch.

the multiplexed sample and hold is memory. The multiplexed sample and hold has the ability to remember the voltage that was applied to the desired output (for a while).

Before proceeding, some background information on sample and hold circuits in general will prove instructive. A functional schematic of a sample and hold circuit is shown in figure 3. The major components of the sample and hold circuit are an electronic input switch, a memory capacitor and a high impedance buffer. The sample and hold circuit works as follows: The voltage to be remembered is applied to the input. The switch is closed, allowing the voltage to be applied to the capacitor. The switch is then opened and the input voltage can now be changed because the output of the sample and hold now reflects the formally applied voltage. This discussion assumes ideal components. There are a number of error sources. The majority revolve around the memory capacitor.

The remembered voltage is stored as an electric charge in the memory capacitor. Because of this any variation of charge with time, ie: current, causes an error in the remembered voltage. This explains the need for a high input impedance in the buffer amplifier so that it doesn't drain away too much charge. A measure of a sample and hold circuit's ability to retain the voltage to within a certain percentage is its hold time. Another problem associated with the memory capacitor is acquisition time. This arises from the fact that charge cannot be delivered instantly to the capacitor. A finite amount of time is needed to deliver enough charge to change the capacitor to the new voltage. Thus we have two design parameters: hold time and acquisition time.

The multiplexed design is not much different from the principles outlined above. The sample switch is merely replaced by our friend, the analog multiplexer. The output to be changed is selected, the strobe then enabled (closing the switch) and after the acquisition time, disabled (opening the switch). The select word and the strobe will

possibly originate from the lower four bits, for an 8 output system, of an output port from the computer.

This type of interface does represent a burden on the computer. The reason for this burden is the very finite hold time of the capacitor. No sample and hold circuit can retain its value forever. The time can be increased by using a larger capacitor, but a limit is reached because a larger capacitor leads to longer acquisition times. Thus, the sample and hold device must be updated periodically if the outputs are to remain accurate. This situation is not unlike that of dynamic memories which are effectively two state sample and hold circuits.

Use of the Multiplexed Digital to Analog System

Obviously this type of interface will require much more computer intervention. The software, though, is not difficult. A possible IO driver is flowcharted in figure 4. For an 8 output system an 8 entry data file is needed to hold the current output values. Periodically (perhaps cued by the interrupt system) with a period less than the hold time a routine is executed to update the eight outputs. The main loop of this routine consists of the following: The value of the output presently being updated is sent to the digital to analog converter interface. The output is then selected and strobed. This action then repeats until all eight outputs have been updated.

What Time Multiplexing Buys You

For the small systems user minimizing hardware is essential. The potential saving of a time multiplexed analog interface is high. The reason for this in general is the reduced hardware. Another not so apparent reason is the ease of expansion.

The most obvious hardware savings occurs in the digital to analog converter. This is because eight channels can be had for the cost of one converter, one 8 to 1 analog multiplexer, eight memory capacitors, eight output buffers, and two output ports. This is contrasted to the eight converters, and the eight output ports from the computer needed by the conventional brute force approach. On the input side the gain is not as obvious. Here we have replaced eight comparators with one comparator and a multiplexer. Both cases require an output port, an input port and a digital to analog converter. A check of prices reveals, though, that eight comparators cost more than one multiplexer.

The clincher is when one considers updating the system to more than eight chan-

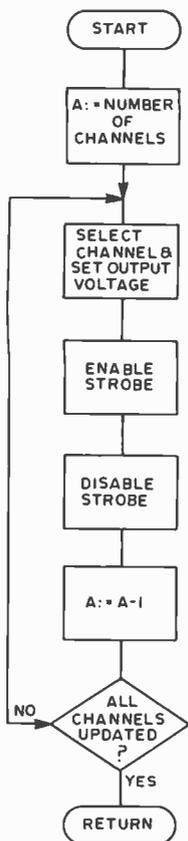


Figure 4: Flowchart of a typical IO driver for a digital to analog system. This program is called periodically to update the value that is being held by the sample and hold circuitry.

nels. The output channel needs only to increase the number of sample and hold devices which is cheaper by factors of eight. The brute force scheme required one more converter and one more output port per additional channel. Table 1 compares the number of components needed for a conventional and a multiplexed system based on 16 channels.

The multiplexed analog to digital conversion process has an even more spectacular success. In the dedicated hardware approach each channel requires one comparator and one bit of input to the computer. The increment for the multiplexed approach is only one multiplexer per eight additions, and one bit of output. Table 2 compares the needed components, based on 16 channels. Thus overall we see that the multiplexed approach offers a multitude of hardware savings.

What Time Multiplexing Costs You

This design technique is typical of many that trade hardware for software. Obviously since we have taken so much out of the hardware, the software and computer efficiency will degrade. It is perhaps a truism that if the interface is designed intelligently these problems can be minimized.

The analog to digital interface suffers the least. The main problem here is the amount of time spent doing the conversion routine. If fast changing inputs or a multitude of moderate inputs are to be converted then the computer is severely loaded. However, many applications only require moderate conversion rates. Foremost of these are interfaces to human operators. Maximum conversion rate needed here is around 100 Hz. Typically, this is around 0.001 Hz to 0.1 Hz. Examples of this are the proportional controls in games and operator set parameters. Control signals in electronic music also fall in these categories. Thus this type of interface can work well in many

GLOSSARY

Acquisition Time: The time required for a sample and hold circuit to change from its previous value to its new value within a prescribed tolerance.

Analog Multiplexer: A solid state device that allows a multitude of connections to be accessed by a common line. The action is like an N position switch.

Comparator: An analog device whose output is logical 1 if the plus input is greater than the minus input and logical 0 if the situation is reversed.

Digital to Analog Converter (DAC): A device whose output analog signal (current typically) is proportional to a digital word at its input.

Component	Dedicated	Multiplexed
Digital to analog converter	16	1
IO ports	16+16 bits	2
8 to 1 multiplexer	0	2

Table 1: Comparison of the amount of hardware that is needed for the direct method of digital to analog converter versus the multiplexed method of interfacing. The table is constructed for an interface consisting of 16 channels.

Component	Dedicated	Multiplexed
Digital to analog converter	1	1
IO ports	3	2
Voltage comparators	16	1
8 to 1 multiplexers	0	2

Table 2: Comparison of the amount of hardware needed for direct methods of analog to digital interfacing versus the multiplexed method of interfacing. The table is constructed for an interface consisting of 16 channels.

typical applications if the rates and the numbers are not excessive.

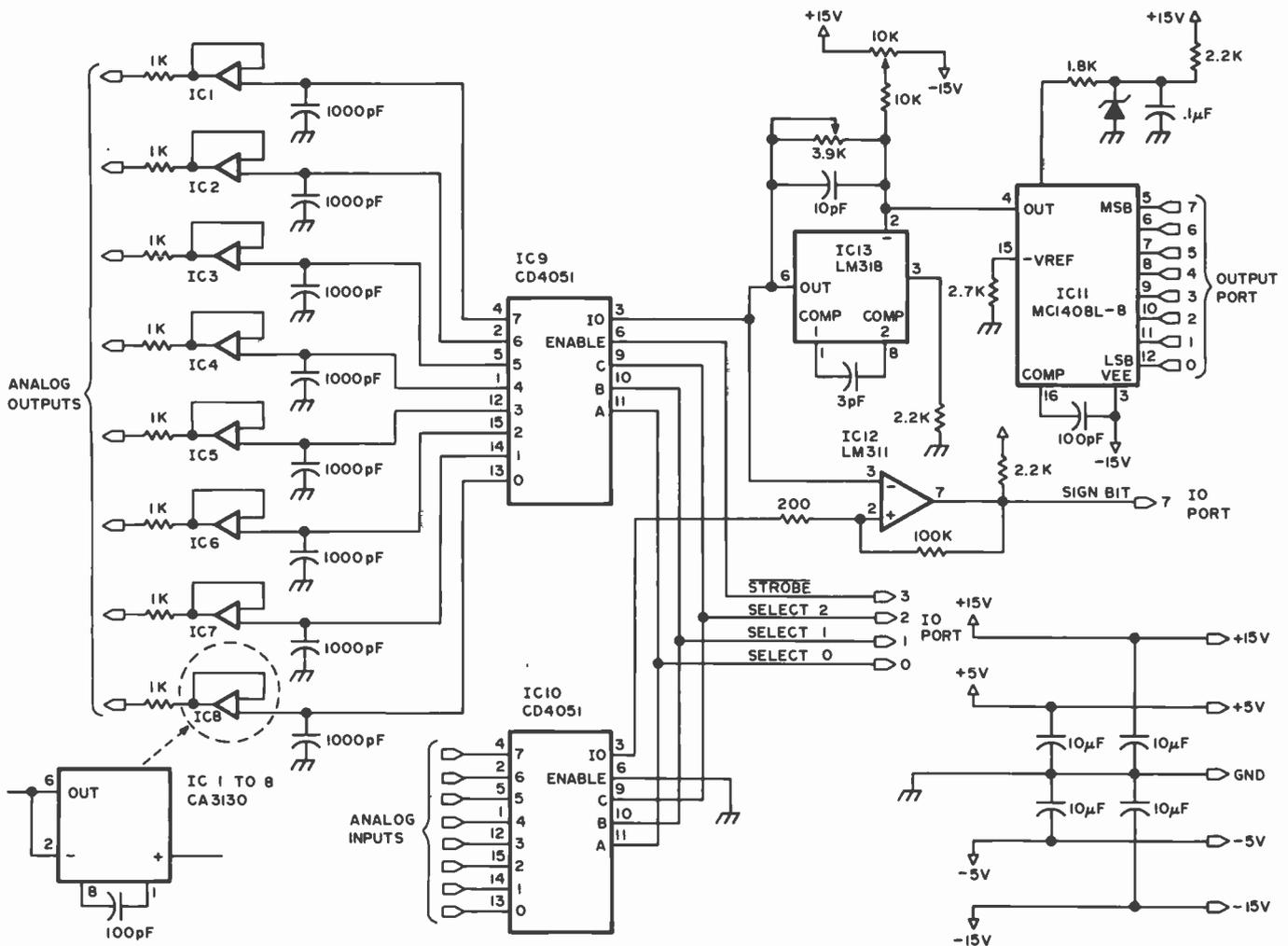
The digital to analog interface suffers from a similar situation. Here the main problems are extremely slow changing outputs or large numbers of fast outputs. Here again an analysis of likely applications reveals that a great number of output signals reside in the frequency spectrum between 1 Hz and 100 Hz. Sample and hold circuitry can be economically designed with hold times in the excess of one second. Also the whole refresh process can be made transparent to the main program if it is done under interrupt control by causing the update routine to be executed at the rate of the fastest output in response to the request of a programmable timer. As with the analog to

High Input Impedance Buffer: A device whose input draws little current from any other devices connected to it. It is important in a sample and hold circuit since currents cause the voltage held in the memory capacitor to discharge.

Hold Time: Amount of time that passes before the output from a sample and hold circuit changes from the originally held value by a prescribed tolerance.

Sample and Hold: Analog memory device which stores a voltage as electrical charge in a capacitor.

Time Multiplexing: Process of combining several measurements for transmission over one signal path. In our case, this signal path is the IO port structure of a processor.



Number	Type	+5 V	GND	-15 V	+15 V	-5 V
1 to 8	CA3130	7				4
9	CD4051	16	8			7
10	CD4051	16	8			7
11	MCI408L-8	13	2			
12	LM311		1	4	8	
13	LM318			4	8	

Figure 5: Schematic of the multiplexed analog to digital and digital to analog interface. The parts were mostly chosen for speed and cost considerations. The integrated circuits are mostly CMOS. All resistances are measured in ohms and all resistors are 1/4 W. Be sure to bypass each power pin with a 0.01 μF capacitor to help eliminate any stray spikes originating from power surges.

digital cases this type of interface works well with a moderate number of medium speed outputs or a multitude of low speed outputs. Note that outputs which can be changed at a rate up to that of the acquisition time (if only one channel is used) can be had by using fewer channels in the output program loop. Outputs that meet these requirements are some automatic testing signals, mechanical devices and control signals for electronic music.

To summarize, this type of interface burdens the computer system. Using it wisely for moderate types of signals does lead to a workable system. Conveniently these types of signals are quite prevalent in interfaces to humans, and many types of equipment.

A Complete Design Example

To illustrate these principles in more concrete terms a complete interface is described in figure 5. The 8 channel interface represents hardware well within the realm of most small systems users. It is optimized for signals from 0.1 Hz to around 100 Hz, though lower and higher rates are possible at the expense of efficiency. This type of interface is useful in interactive games, testing of equipment and electronic music. Parts used are neither exotic nor expensive.

Figure 5 contains the complete circuit. The following comments on component selection are in order. The digital to analog converter was chosen for cost reasons. However, almost any state of the art current output converter can be used. Note that a

multiplying converter can provide for scaling of the output by a voltage, possibly from another interface. The LM318 operational amplifier was chosen to be the current to voltage converter to minimize response time of the DAC. As such, only an operational amplifier of similar speed should be substituted unless you can tolerate slower response. Also care must be taken to isolate the amplifier from stray signals, or it could become unstable and oscillate. The multiplexer chosen for both converters is one of the CMOS variety. In order to meet the specifications on this device and give an easy design, the voltages of this system are restricted to ± 5 V. This allows for adequate range for most 8 bit applications. 10 V full scale means one bit is 39.1 mV. This can be scaled down if needed.

The sample and hold capacitor was chosen to satisfy the acquisition and hold time requirements. Changing its value can tailor the system to individual needs. Always use polystyrene capacitors since their characteristics are essential to a good interface. I used the new CMOS operational amplifiers for the buffers because of their high impedance and low cost. Notice that they operate on the same power supplies as the analog switches.

The comparator was chosen for its low cost and speed. Similar devices could be substituted. The necessary power supplies are ± 15 V and ± 5 V. The ± 15 V could be reduced to ± 12 V if convenient.

The software for such an interface is not difficult. Roger Frank's article and figure 4 contain the basics. A complete routine written for an MOS Technology 6502 based system is shown in listing 1.

Summary

A multichannel analog interface can be designed with a minimum of hardware if a

Listing 1: Typical program written for a 6502 based system to update eight digital to analog conversion channels. The program sequentially addresses each channel, outputs the desired voltage to be held, disables the channel, and steps to the next channel. This is done once for each time that the program is called. This program could be set up as an interrupt handler which responds to a clock strobing an interrupt line.

Label	Op	Operand	Commentary
UPDATE	LDX	#07	initialize pointer;
LOOP	LDA,zpa,X	BUF	get next byte for output;
	STA	DAC	output byte;
	TXA		accumulator:=pointer;
	STA	CONT	select channel and enable sample and hold;
	ORA	04	turn sample and hold strobe off;
	STA	CONT	turn selected sample and hold off;
	DEX		pointer:=pointer-1;
	BPL	LOOP	if pointer >=0 then go to LOOP;
	RTS		else return from subroutine;

Data Definitions

BUF: A string of output bytes
 DAC: Address of DAC output port
 CONT: Address of control port

few software vs hardware trade offs are made. Though the multiplexed approach does impose some software burdens, for most applications the variation of the outputs and inputs is slow enough to make this type of interface transparent under the interrupt system. This type of interface should make many real world applications possible to the limited budgets of most experimenters.■

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3. Kraul, "Analog Interfaces for Microprocessor Systems," *Electronotes, Newsletter of the Musical Engineering Group*, volume 8, number 63, March 1976, pages 11 to 14.

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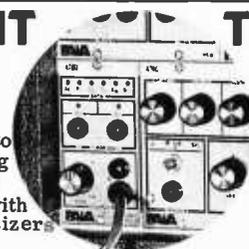
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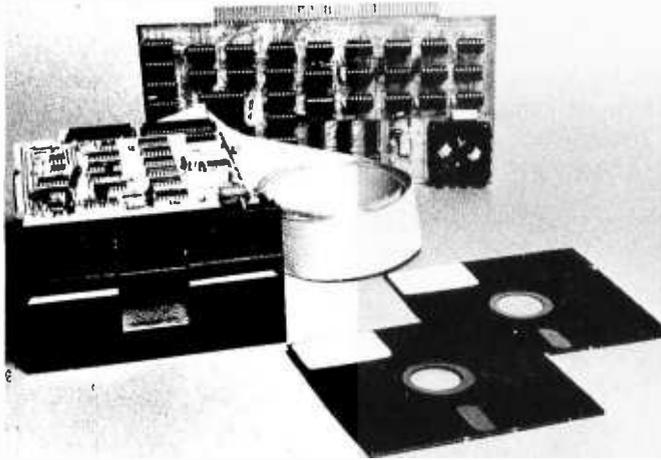
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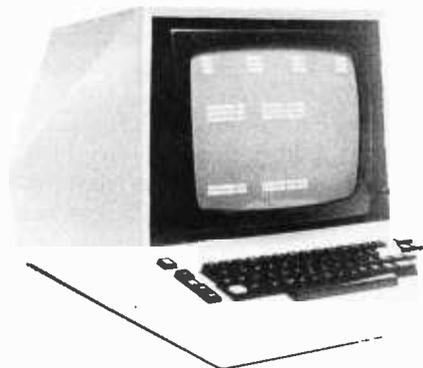
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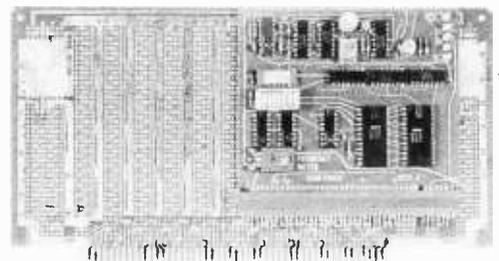


Volker-Craig Ltd, 266 Marsland Dr, Waterloo, Ontario CANADA N2J 3Z1, has introduced a new data terminal intended for use in small business systems, timesharing and other small computer end user applications. The features of this terminal include RS232C interface at rates from 110 to 9600 bps, a 1920 character display (24 lines of 80 characters), 12 inch CRT display, composite video output for extra slave monitors, XY cursor addressing by computer or operator, 64 key ASCII keyboard with tactile feedback and automatic repeat, and an optional separate numeric data entry keypad. Options include upper and lower case ASCII keyboard and display, switched serial interface, parallel input and output interfaces, and custom keyboard character fonts such as APL or French. The display is the model VC303A, and the price is \$1595 in quantities of one, \$995 to distributors and OEM quantity buyers. ■

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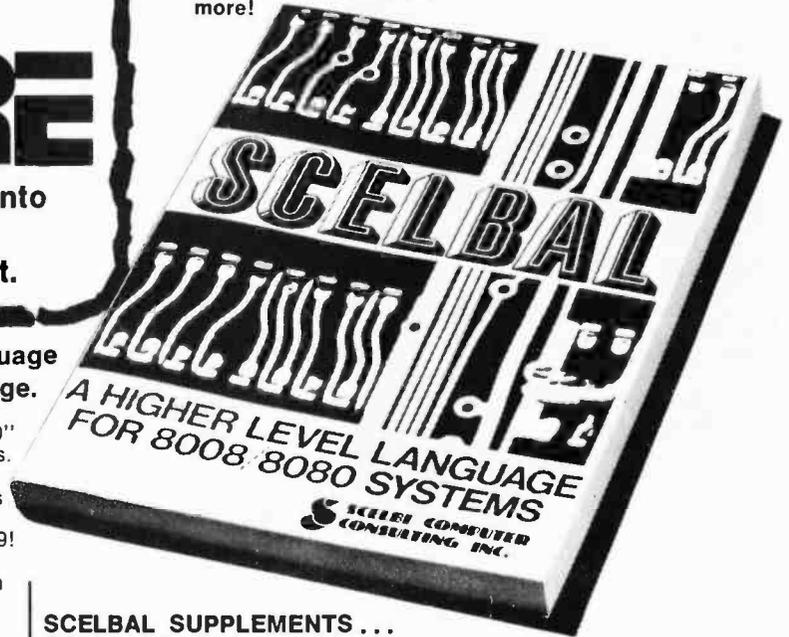
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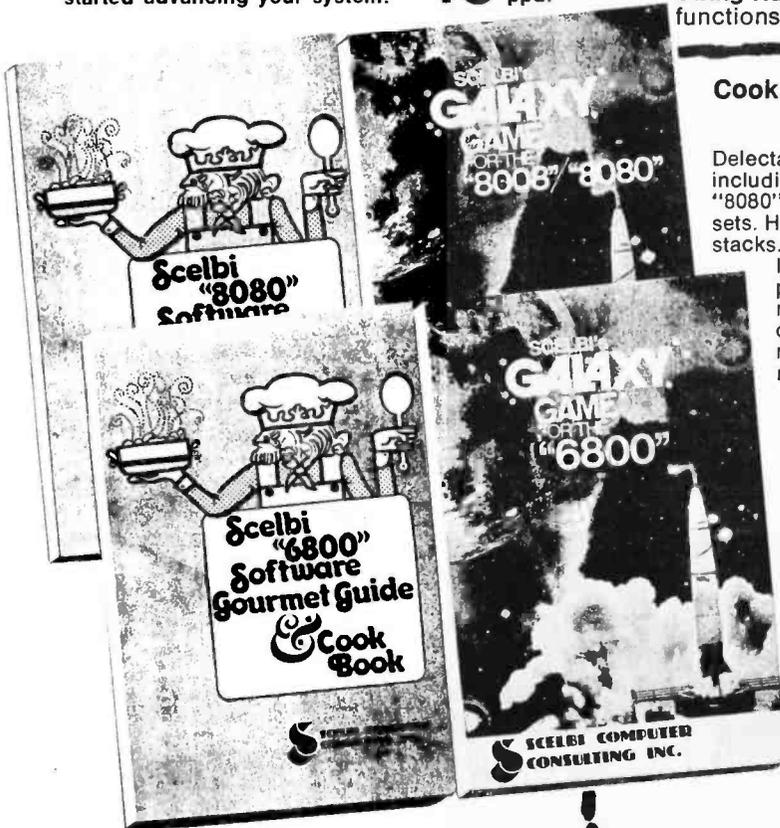
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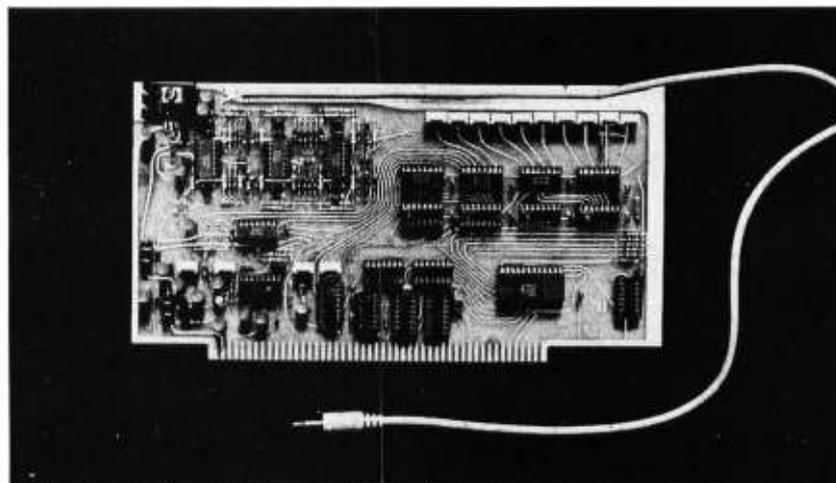
125, Wayland MA 01778, stopped by BYTE's offices in early February, toting an Altair 8800 under his arm and an interesting peripheral which his firm has designed and is marketing. The peripheral is the AIM-1005 frequency meter, demonstrated to us mounted upon a Vector prototyping card for the Altair bus. (This was a demonstration setup; his firm will soon have a \$30 card with all the connections to the Altair bus made for piggyback mounting of the frequency meter on a permanent basis.) What this \$178 product does for its user is provide a programmable frequency meter with 13 bits of precision, and 11 different time base ranges allowing measurement over time periods from 10 μ s to one hour. The input logic has scaling counters with an upper limit of 25 MHz, so it is possible to make a quite useful general laboratory frequency meter with outputs on a computer terminal by simply driving this peripheral with a simple assembly language or BASIC program. The device is interfaced through memory address space, decoding the high order four bits of the 8080's 16 bit address bus. ■

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A Synthesizer Example . . .

AI Cybernetic Systems, POB 4691, University Park NM 88003, has sent along some new information on their Model 1000 Speech Synthesizer, which was first described in Wirt Atmar's article in August 1976 BYTE, page 26. The picture here is the production version of this device, which is an analog model of the human vocal tract, digitally programmed with commands that correspond to American English phonemes. Since the device works with coded phonemes, the maximum information transfer rate for speech which is required is about 50 bps. The device plugs directly into the Altair bus and cost is \$325, with delivery from stock.

A significant and interesting bit of literature which came with this picture was "Programming Example 1," an



A New Brochure



E & L Instruments, 61 First St, Derby CT 06418, has just released a new 8 page brochure describing the MMD-1 education and development micro-computer and its optional accessories. Copies of the MMD-1 brochure are available from the company, as well as its representatives and authorized dealers. ■

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Altair BASIC program with an 8080 machine language subroutine for details at a low level. The program is an implementation of the well-known Lunar Lander game, but with the novel twist that the "ship's computer" tells you your present height from the surface of the planet as you land. It will run on any Altair 8800 equipped with 12 K of memory. So, what you do is plug in the Model 1000, read in BASIC, type in this program and set up its machine language support routine, then proceed to use a new mode of interaction with the computer system. ■

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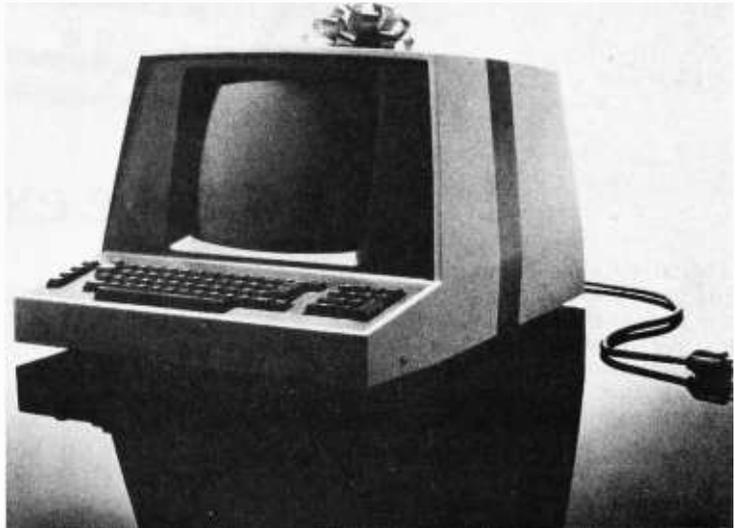
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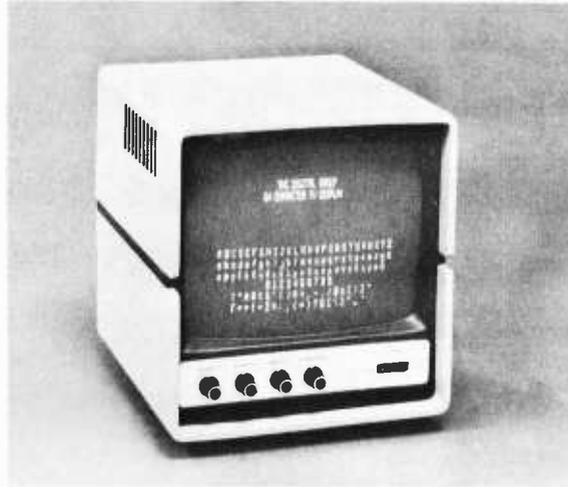
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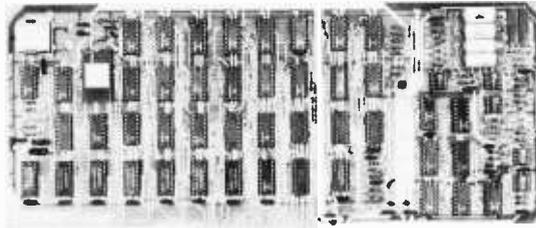
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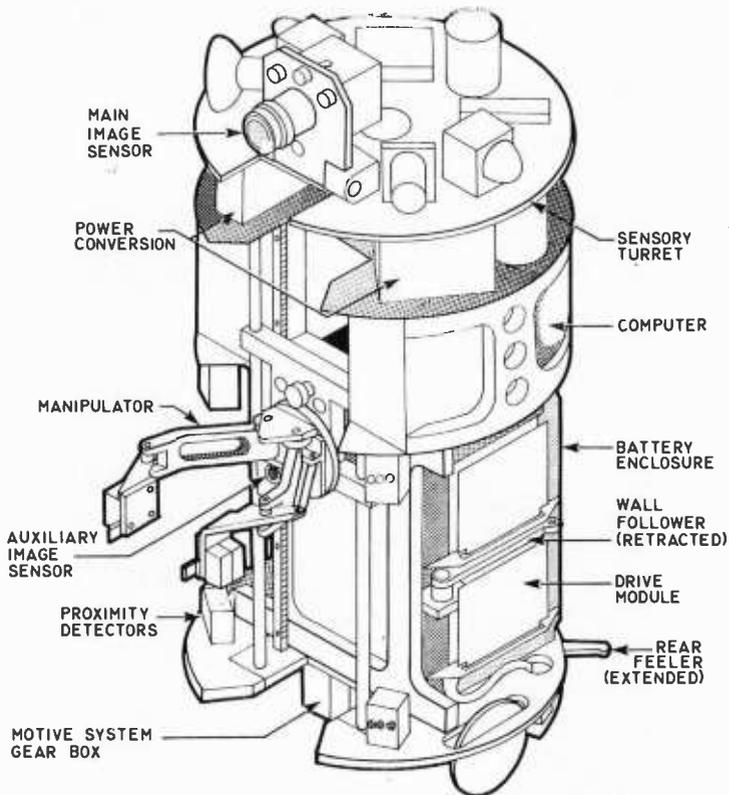


Figure 1: The robot Newt as it will appear when completed. There are three main subsystems shown in this diagram which was supplied by the author: motive, manipulator and sensory turret. The entire machine is controlled by an on board microcomputer. The motive power subsystem is built around two precision drive wheels actuated by stepper motors. The manipulator, a simple hand capable of grasping, lifting and rotating, is also actuated by stepper motors and includes various sensors. The sensory turret at the top is a platform which includes a main image sensor, which can be tilted to look straight down or up at a slight angle using a stepping motor; other sensors are shown in outline form, and are a subject for future experimentation.

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In the late 1930s, a young man named Rossum began manufacturing industrial robots in a small factory on the outskirts of Prague. This venture was immediately successful and would have virtually guaranteed a second industrial revolution had it not been for a singular tragic circumstance: The robot workers became irrational and revolted. They turned on their masters and burned the factory to the ground.

Fortunately, the preceding scenario is only a work of fiction by the Czechoslovakian writer K Čapek. Since Čapek's coinage of the word "robot" in his 1923 play R U R it has been the subject of a great many works of science fiction, including Isaac Asimov's *I, Robot* and the movies

"Forbidden Planet," "Gog," "Silent Running," "Westworld" and others.

Alas, the imaginative stories of the science fiction writers have far outstripped the efforts of the robot engineers. Progress in building real robot devices has been painfully slow over the past several decades, although a few individuals and small groups have produced some very interesting results.

It is not my purpose here to review these works, but rather to describe a project with which I have been personally involved for some time: the construction of a freely moving robot vehicle which will be capable of interacting with its environment in a rational way, and managing its own survival. The work is being done by a small group of people in the Duane Physical Laboratory of the University of Colorado. The project is being financed through personal funds, and in this sense qualifies as an "amateur" undertaking.

What are the requirements for such a robot? It must be able to explore the environment in some orderly manner, measure the attributes of objects and obstacles encountered, classify them according to some scheme, and incorporate them in its evolving internal world model. The world model must have a logical structure which allows modifications to be easily made; it must be compact with respect to memory space, and it must use a design which can be consulted in some reasonable way. The robot must be able to manipulate the world model (cognition), derive informed decisions from it, and carry out these decisions in physical action to achieve broadly defined goals.

Figure 1 is a general view of the robot as it will appear when completed. Practically speaking, it resembles less a mechanical man than, say, a shop type vacuum cleaner. The machine is cylindrical, about 36 cm (14 inches) in diameter and 76 cm (30 inches) in height, weighing approximately 27 kg (60 pounds). All rigid mechanisms lie within the cylindrical boundary or can be retracted within the boundary if necessary. This design greatly reduces the number and diversity of senses required. The robot will be fitted with a smooth cylindrical skin (not shown), removable in sections for easy inspection and maintenance. Modular construction is used in the robot wherever possible. There are three main subsystems: motive, manipulator and sensory turret. These are presided over by an 8080 based microcomputer with 8 K bytes of EROM and 24 K bytes of programmable memory. The entire system is powered by a 6 V, 84 amp-hour storage battery.

How the Robot Gets Around

The robot is given locomotion by two main diametrically opposed drive wheels. A third (unpowered) castoring wheel is located at the rear. This motive geometry has been successfully employed in several other robot vehicles constructed by other groups, and has been in use in the author's robot research since 1957. The vehicle's center of gravity is located well aft due to the placement of the battery and other heavy components, obviating the need for a front wheel. The wheels are constructed of aluminum with neoprene O-ring tires. The main wheels are driven by stepping motors through precision 3:1 gearing. A single command step rotates a wheel 0.6 degrees, so there is a total of 600 steps per revolution. If both wheels are moved in the same direction, the vehicle travels forwards or back-

wards in a straight line. If the wheels are moved in opposite directions, the vehicle executes a perfect rotation about its vertical axis. By stepping the wheels at different rates, a circular trajectory is approximated having a radius which can range from zero to infinity. The hardware and software necessary to drive the stepping motors are discussed in some detail later in this article, since they have very general application.

The robot navigates principally by open loop dead reckoning. That is, it depends largely on the precise control of the wheels during acceleration, deceleration, and constant speed motion to achieve accurate positioning of the vehicle. The robot's position and azimuth relative to a fixed origin are computed at the end of each motion segment by counting stepping motor increments and using trigonometry. The precision attainable is limited principally by wheel slippage, unequal wheel diameters, nonplanar floors, round-off errors, and step quantization. In practice, all these errors are small for short distance movements.

The robot's excellent open loop performance makes it unnecessary to have continuous closed loop servo systems, such as have been extensively employed in other robots. For example, consider the problem of having a robot view a small object on the floor a short distance away, and then go pick up the object. One approach would be to have the robot continuously view the object as it moves towards it, adjusting its motion in a continuous way to converge on the object and pick it up. In this approach, a rather crude motive system would suffice, since errors are always nulled. A heavy load, however, is placed on the sensor-computer system. An alternative approach, the one followed here, is to have the robot view the object once, and then compute exactly how to move to the object and pick it up. When the computation is completed, the machine simply carries out the proper motions in a blind fashion. There is a greatly reduced strain on the sensor-computer system, at the expense of having to build rather precise motive machinery. Of course, this approach assumes the relevant environment will remain fixed for the duration of the task.

In this general spirit, by using high precision open loop movements throughout the robot, only intermittent feedback through the senses is required to close the control loop. In this way, the overall complexity of the robot can be kept at a reasonable level, allowing the required computations to fall within the abilities of an on board microcomputer.

Epistemological Engineering

With artificial intelligence, robotics and applied cybernetics coming of age through the recent progress in the fabrication of computer systems, it is quite likely that the branch of philosophy called epistemology will have a much more explicit role in technology during the coming years. Epistemology is the study of nature and grounds of knowledge; understanding of epistemology is crucial to any attempt to realistically implement artificial intelligence systems. So in a future world of cognitive automata and advanced information systems, we may indeed find the new specialist who is the "epistemological engineer."

Scanning the Environment with Electronic Senses

The sensory turret (see figure 1) provides a general platform on which to mount various senses, some of which might be quite experimental and temporary in nature. It can pan a full 360 degrees, and a small section containing the main image sensor can tilt from approximately 30 degrees above horizontal to 90 degrees below horizontal. Both of these motions are controlled by stepping motors. The main image sensor has a motorized focus control. The geometry of the "hand-eye" system allows orthogonal views of objects held in the manipulator jaws by using both image sensors under conditions of controlled focus and lighting.

Each image sensor is a 32 by 32 element integrated array of photodiodes on a silicon chip measuring approximately 4 mm (0.1 inch) square. The array acts like a 1024 bit memory. Each element is precharged to a fixed voltage; then at some later time the voltage of each element is read out, decreased in proportion to the amount of light which has fallen on it. Several milliseconds are required to digitize the image. Up to 16 levels of gray can be discerned, which means that 512 bytes of computer memory are sufficient for a single image. Once the image is obtained, it is analyzed by appropriate software. The extremely small size of these solid state image sensors, and the simplicity of their associated electronics, make them ideal for robot use. Higher resolution devices such as charge coupled sensors and miniature television cameras are available, but their cost and complexity make them unattractive for such applications. Besides, the amount of data generated would be too unwieldy to handle with a microcomputer.

In addition to the main image sensor, several simple phototransistor light sensors are mounted on the tilting platform of the sensory turret. These enable the robot to locate and track point sources of light.

Several other proposed senses are to be mounted on the turret. These are shown in schematic form in figure 1. To the left of the image sensor is shown an ultrasonic ranging transmitter and receiver which should be extremely useful for finding the range to walls at distances as great as 10 meters (33 feet). The idea is to transmit short bursts of 40 kHz sound and measure the time required for an echo to be received in order to compute the distance. Much of the necessary circuitry is available in compact integrated form.

Just to the right of the image sensor is shown a long wavelength infrared radiation detector with which it might be possible to

locate sources of heat such as dogs, cats and humans.

Located on an axis perpendicular to the main image sensor are shown two microphones which form part of a sound location system. The intent is for each sensor to acquire a short sample of sound. The computer then attempts to find a phase relationship between them, thereby locating the direction of the source.

Many other possible senses can be imagined; the sensory turret is intended to provide a versatile platform and interface for experimenting with them. For example, using a helium neon laser, or perhaps a compact metal vapor laser, it may be possible to provide an optical range finding system with high resolution if a suitable detector can be found. Eventually it is hoped to provide limited forms of speech synthesis and phrase recognition (using off line electronics with an analog radio link). This would make it possible to give general spoken commands to the robot in contextual surroundings, and have verbal feedback to insure that the commands were being properly interpreted.

Sharing some space with the sensory turret system, and directly below the platform in figure 1, is the necessary power conversion electronics. These modules convert the battery voltage to ± 12 V and +5 V (regulated) for use by the electronics, and also house the battery recharging circuitry.

Manipulating Objects in the Environment

The number of tasks which a simple robot can do is greatly increased if it has some sort of hand with which to grasp and manipulate objects in the environment. The human arm and hand system, with its 27 degrees of freedom, is a marvelously versatile mechanism. Large industrial manipulators are fairly complicated and have six or seven degrees of freedom. For practical reasons, the present system is limited to a mere three degrees of freedom which, when combined with the two degrees of freedom of the motive system, should be sufficient for carrying out simple tasks.

The manipulator is able to grasp objects, move up and down along the front of the robot by means of a rack and pinion drive, and to rotate about a horizontal axis. All three motions are controlled by stepping motors. The manipulator has a parallelogram geometry which permits the jaw faces to remain parallel regardless of their separation. This feature simplifies the problem of picking up objects of varying sizes. When not in use, the jaws open wide enough to bring all

parts of the manipulator within the cylindrical boundary of the robot, where it is out of the way and does not cause problems when the robot turns and maneuvers in tight places.

An auxiliary 32 by 32 element image sensor with fixed focus lens is mounted directly in the manipulator assembly, providing a view of whatever object is between the jaws.

In addition to mechanical force sensors in the jaw faces, there are several infrared LEDs in one face with opposing phototransistors in the other face. These LED phototransistor pairs define beams of light between the two jaw faces. By scanning the manipulator up and down, and moving the entire vehicle forwards and backwards, these beams enable the robot to measure the height and depth of simple objects resting on the floor between the jaw faces, as well as to determine when objects are correctly positioned for picking up. The widths of objects are measured directly by counting the number of stepping motor pulses required to close the jaws on the object.

A tactile sensor is mounted on the front of each jaw to enable the robot to sense and locate large obstacles such as walls. The robot can align itself accurately either parallel or perpendicular to the wall, after performing a simple series of maneuvers. This action is convenient for many of the tasks the robot might have to carry out. For example, a simple strategy of maneuvers allows the robot to "square itself off" from a corner in a room, thereby defining a precise origin for its navigational coordinates.

The tactile sensors double as electrical contacts used for plugging into ordinary wall outlets for the purpose of recharging the robot's battery (more about this later).

All electrical power and signals to and from the manipulator assembly are sent through folding ribbon cables which are not shown in figure 1.

Auxiliary Senses

In addition to the senses mounted on the manipulator and sensory turret assemblies, there are several auxiliary senses illustrated in figure 1. A pair of retractable feelers are provided in the rear of the vehicle and serve much the same purpose as the tactile sensors on the manipulator jaws. These rear feelers enable the robot to detect and square up from walls and corners if the manipulator is occupied with holding some object.

In addition to the rear feelers, a pair of retractable wall-follower feelers located on the sides of the vehicle enable the robot to

travel parallel to a wall by measuring the distance from itself to the wall. The rear feelers and the wall-follower feelers are both activated by double coil latching solenoids and contain "microswitches" consisting of movable masks and infrared LED-photo-transistor pairs.

For all of the robot's varied and specialized senses we have discussed so far, there is nothing to prevent the robot from accidentally running at high speed into a wall. To prevent just such a mishap, and to provide a "soft" broad area "sense of touch," the robot is equipped with a number of proximity detectors, several of which are shown in schematic form in figure 1. These detectors work on a simple, but elegant, principle. Infrared LEDs, amplitude modulated by a 20 kHz signal, send broad beams of light out from the robot. If a wall is nearby, some of this light is reflected back into phototransistor sensors located in the same module. The received signal is amplified and sent to a phase sensitive detector locked to the outgoing signal. The computer is notified if the signal exceeds some programmable threshold. These proximity detectors are quite insensitive to ambient light conditions, not very sensitive to the color or texture of the reflecting surface, and have useful ranges extending from several centimeters to perhaps one meter. With these detectors, the robot is free to travel at fairly high speed until a wall or other obstacle is encountered, and can then slow down and investigate it with caution.

It should be mentioned that all of the senses interact with the processor through a vectored priority interrupt system.

Robot Psychology

Up to the present time, the environment of the robot has been intentionally restricted. The robot is permitted to roam freely within a large "playpen" constructed of plywood. As various sensory functions are added, and the robot nears completion, simple objects such as blocks of wood, and larger obstacles constructed from plywood will be introduced, and restrictions on the environment will be gradually relaxed. While this is happening, the complexity of the software will greatly increase. It is expected to take several more years of effort before the robot reaches a form which can be considered "finished." Whether the robot will ever be able to cope successfully with a "general" environment, such as a typical research laboratory or home living room, is certainly an open question.

The software will be developed as a hierarchy of modules. The bottom strata of

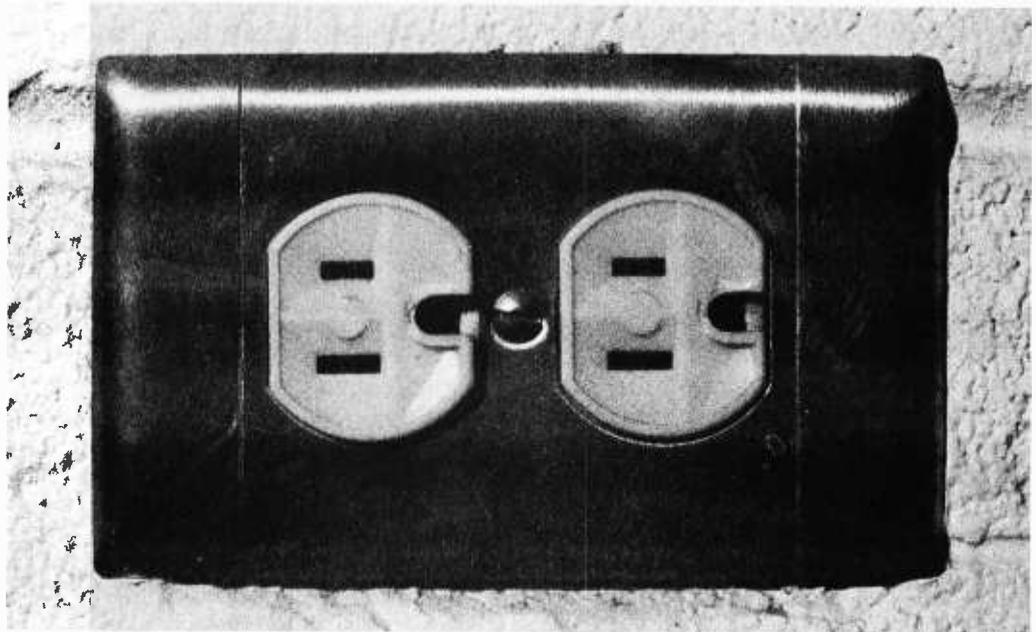


Photo 1: The feeding trough. This is a photograph of an ordinary wall outlet with a white concrete block wall as a background. The robot must be able to seek out and plug into outlets such as this in order to periodically replenish its energy supply.

this hierarchy will be occupied by hardware oriented routines for motor control, interrupt servicing, sensory data acquisition and the like. These routines will be relegated to EROM, where they will reside as powerful appendages for the higher level software. In the highest reaches of the hierarchy will be the planning algorithms: heuristic programming which will take into account goals, subgoals, behavior, the structure of priorities, and what may generally be called the "psychology" of the robot.

If the robot is turned loose in a strange room, its immediate behavior will be to begin exploring the room in some systematic way, locating walls and the boundaries of objects with respect to a specific origin. As new features of the environment are encountered, they will be incorporated into the world model. Particular interest will be shown in the positions and heights of recognizable wall outlets, since these are critical for survival. As the model of the room nears completion, the original intense curiosity will subside, and will be replaced by an attitude of playfulness. The world model will be consulted to find out which objects (eg: wooden blocks) are small enough to be manipulated. The robot may choose to group together objects with similar characteristics, build simple structures by stacking blocks and so forth. When the energy supply nears exhaustion, the robot will interrupt its play, head directly for a convenient wall outlet, plug in, and recharge its battery. The recharging process might take several hours, during which all motors will be turned off, and all but one 4 K byte memory module will be powered down. After recharging, the

robot will unplug itself and continue on its way. When the robot eventually "tires" of its environment, it will perhaps leave the room, wander down the hallway, and look for other rooms to explore. An interesting experiment would be to observe how long it takes for the machine to recognize a room in which it is arbitrarily placed, but has previously explored. In addition to observing general behavior, one can give the robot general tasks to perform, such as picking up all small objects in a cluttered room and placing them in a box in the corner. (Mothers with small children, take note!)

Many of the computing requirements for the robot will exceed the capabilities of its on board microcomputer. For example, the analysis of complicated scenes viewed by the image sensors can, at present, be done only by a very fast-computer with lots of memory. For this reason it is planned to equip the robot with a duplex radio telemetry link to a "black box" which connects with a telephone. The robot will then be able to initiate telephone calls to a large timeshared computer, and be able to communicate with it via the 2400 bps telephone lines. Control routines in the on board microcomputer will invoke large analysis programs stored on permanent disk files attached to the timeshared computer.

A visual scene can be transmitted in less than a second, and the large computer can extract features of interest and send them back to the robot in several seconds. It will also be useful to take advantage of the off line disk files to store large data bases such as portions of the world model. When the robot has finished its transactions with the

large computer, it simply "hangs up the phone" and goes about its business, perhaps reinitiating the hookup at a later time. It is important to point out here that in no sense is the large computer "in control" of the robot. The robot is simply using the services of the large computer to perform calculations that are too involved or too lengthy to do itself, much as you or I would use a computer to solve some problem. Also, it should be noted that if the large timeshared computer is busy, or if the computations are long, the robot is free to continue with other tasks until the requested analysis can be performed.

This usage of an off line timeshared computer will permit access to a large amount of general robot planning software written in high level languages by other groups. The telemetry hookup will, as a side benefit, enable the robot to make calls to human researchers (hopefully not in the middle of the night) to report malfunctions or unusual conditions in the environment by means of coded audio signals. (The robot night watchman?)

Finding Energy Sources

It may be necessary for the robot to survive for weeks at a time without supervision or interruption of its operation. To do so, it must manage its own energy supply. The battery voltage and current are sensed periodically and converted to digital form by a simple software driven analog to digital conversion system. This information, along with the battery's internal temperature (sensed with a thermistor probe), allows the state of charge to be determined. When this reaches some minimum level, the robot is obliged to renew its supply of energy by finding a wall outlet and plugging in.

From the outset, it must be said that the robot will depend heavily on preprogrammed strategies to accomplish this task. A standard wall outlet mounted on a white concrete block wall is shown in photo 1. The cover plate is of brushed stainless steel, and the receptacles are made of light colored molded plastic. From a distance of perhaps 10 meters (33 feet), with the turret image sensor focused on infinity, the outlet appears as a small spot. Any such spot, identifiable by the software as a few dark picture elements surrounded by a light field, is worth further investigation. If several spots are visible, a single one will be selected for closer examination.

Using the motive system and the sensory turret in combination, the direction and distance to the spot is then computed by triangulation. After this is done, the robot

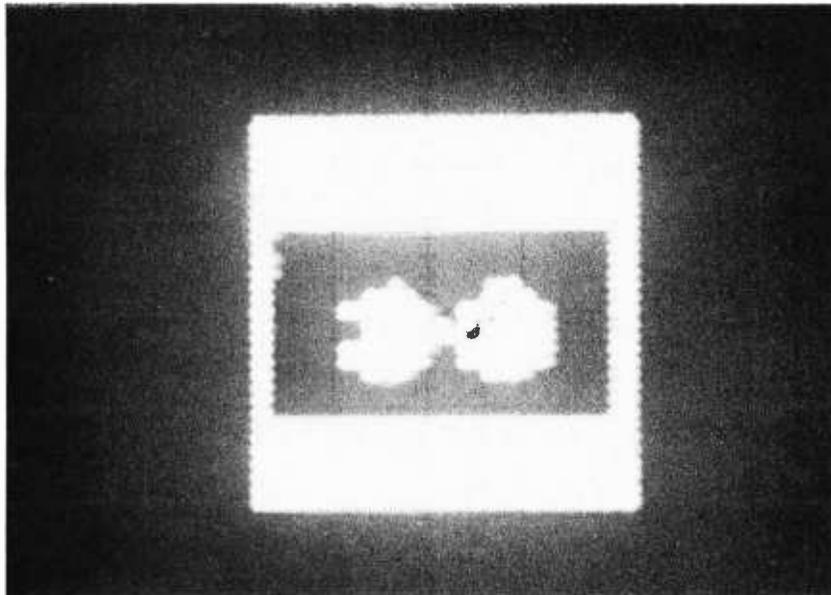


Photo 2: Here is an oscilloscope display of the wall outlet shown in photo 1, as viewed by an image sensor. There are a total of 1024 picture elements in this sensor, each with 16 levels of gray perceptible to the computer. The picture has 32 columns and 32 rows of elements. The outlet appears as a dark rectangle with a light background. The two receptacles can just be resolved at this range.

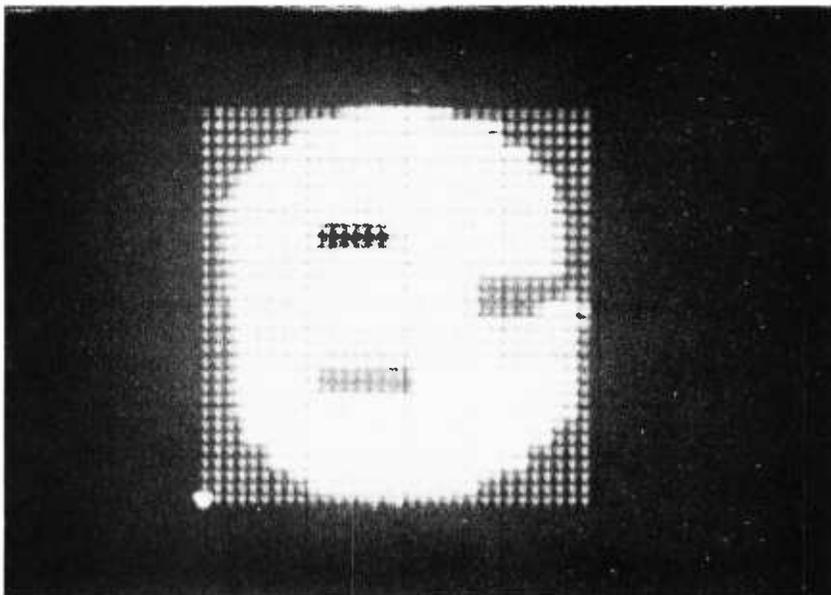


Photo 3: As the robot moves closer to the wall outlet, more detailed features of the socket become apparent. This is an oscilloscope display of an individual receptacle, as viewed at close range by the image sensor. By applying pattern recognition techniques to images such as this and the image in photo 2, the robot can recognize outlets and determine the exact position and height of the electrical contacts, prior to its final maneuver to plug itself into the wall.

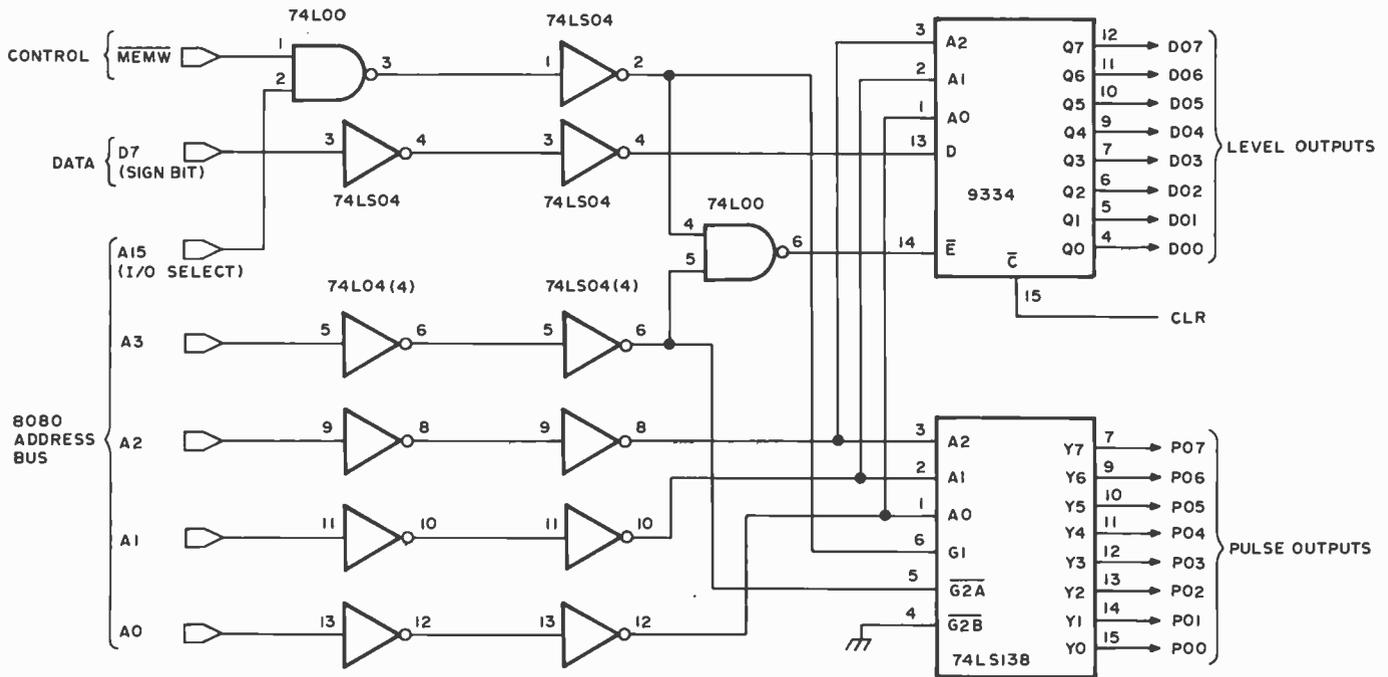


Figure 2: Output decoding logic used to create discrete (level) outputs and pulse outputs. This circuitry connects to an 8080 data bus through memory address space. The outputs are used to drive the various lines of stepper motor interfaces such as those shown in figure 3. Each stepper motor individually uses one discrete level output (for forward or reverse direction control) and one pulse output (to initiate one step sequence). Two discrete level outputs are also used to control the running versus standby, or off, status of the motors.

drives to the approximate position of the spot, and with LED proximity sensors and the tactile sensors on the manipulator, it attempts to sense and recognize the existence of a wall, which is hopefully associated with the spot. If this is the case, the robot moves out perpendicular to the wall for a fixed distance, sets the image sensor focus accordingly, and rescans for the spot which hopefully should now begin to have the shape of a wall outlet from this short range. By a series of small maneuvers, the robot positions itself exactly in front of the outlet at a precise distance away. At this time, the manipulator begins a vertical motion until the outlet is centered in the field of view of the auxiliary image sensor. The computer now has a straight on, in focus view such as that seen in photo 2.

At this point, fairly simple pattern recognizing algorithms are used to analyze the shape, size and topology of the image (feature extraction). These features are compared with known properties of wall outlets. If enough features match within certain error bounds, the recognition is successful. (In this respect, it is much easier to attempt to recognize a specific object than to try to analyze a general scene.) If the recognition is not successful, several more attempts are made at slightly different distances to over-

come the granularity effects of the sensor. If a recognition is still not successful, the robot heads off to explore other spots. Assuming success at this point, the robot moves in for a high resolution image of a single receptacle (see photo 3). Again, the image is analyzed, and further recognition tests are made. If everything is okay, the exact height and distance to the electrical contacts are determined, and the manipulator jaws close to the separation appropriate to the space between the two contacts. An attempt is now made to plug in, with the computer monitoring the voltage between the two plug prongs on the manipulator. Small searching motions are made until contact is established, whereupon the jaws open slightly to make good electrical connections. The computer then directs the recharging operation until the correct charge is reached. It is expected that the robot will normally be able to go about six hours between recharging operations.

Stepping Motor Drives

In order to execute all of the motions required of it, the robot must have precise control of the motors. This is accomplished by means of the stepping motor drive modules which constitute the interface between the on board computer and the stepping motors. There is a total of seven

identical drives: two for the motive system (one for each wheel), three for the manipulator (grasp, lift and rotate), and two for the sensory turret (pan and tilt).

Figure 2 illustrates the logic used to provide both pulse and discrete (level) outputs from the computer. For simplicity, only eight pulse outputs and eight discrete outputs are shown. The scheme can be expanded to many more outputs, with the addition of suitable logic. Low power TTL provides the connection to the address, data and control buses of the microprocessor. Lines A₀ to A₃ select one of 16 possible outputs, with A₁₅ acting as an input output select line. The sign bit, D₇, is used to signify whether a one or a zero will be written into the selected discrete output line. Generation of the negative going output pulses and strobing of the data are done by the $\overline{\text{MEMW}}$ pulse. A 9334 addressable latch and 74LS138 binary to octal decoder provide the outputs to the system. If desired, one pulse output can be wired to the CLR line of the 9334 latch for simultaneous resetting of all the discrete outputs.

Figure 3 shows an individual stepping motor drive circuit which is sufficiently simple and general to permit many other non-robot applications where precise computer controlled motion is required. The circuit generates the 4 phase pulse sequence of high currents necessary to operate a Slo-Syn™ bifilar type stepping motor manufactured by the Superior Electric Company. Inputs to the stepping motor circuit are connected as desired to the discrete and pulse outputs shown in figure 2. The forward or reverse input ($\overline{\text{FR}}$) determines whether the motor shaft turns clockwise ($\overline{\text{FR}} = 1$) or counter-clockwise ($\overline{\text{FR}} = 0$) as viewed from the shaft end of the motor. The forward limit input ($\overline{\text{FL}}$) prevents the motor from moving forwards (clockwise) if $\overline{\text{FL}} = 0$. The reverse limit input ($\overline{\text{RL}}$) prevents the motor from moving backwards if $\overline{\text{RL}} = 0$. These inputs are generally not controlled by the computer, but are wired to simple limit switches to prevent excessive motion of the motor in a particular direction. In the case of the robot, these "reflex inputs" are independent to allow the computer to change the direction of motion after a mechanical limit has been reached, and to establish "zero point" settings for the various possible motions. The pulse input ($\overline{\text{P}}$) triggers on a negative going transition and causes the motor to advance one step (1.8 degrees) in the direction specified by the $\overline{\text{FR}}$ input. Thus, 200 pulses on the $\overline{\text{P}}$ input cause the motor shaft to rotate exactly one revolution. The maximum pulse rate is of the order of several hundred pulses per second without error, but this

depends strongly on the motor size and the driven load. The clear input ($\overline{\text{CLR}}$), when momentarily brought low, resets the counting flip flops to zero in case it is necessary to establish a standard shaft position when power is turned on. The off input ($\overline{\text{OFF}}$), when brought low, removes all current from the motor windings, and frees the shaft except for a small holding torque caused by the permanent magnets. If $\overline{\text{OFF}} = 0$ and standby (SBY) is brought high (SBY = 1), the shaft position is maintained, but at greatly reduced current. The off and standby inputs are critical for application in the robot, since they permit substantial power savings. In figure 3, series 7400 logic can be substituted for the 74L and 74LS series if desired, but at the expense of extra power. Resistance values in the drive circuits are typical, and might need to be optimized for a particular motor. The diodes in the emitter circuit of the 2N3055s should have a high surge rating, and moderate heat sinking for 2N3055s is advised.

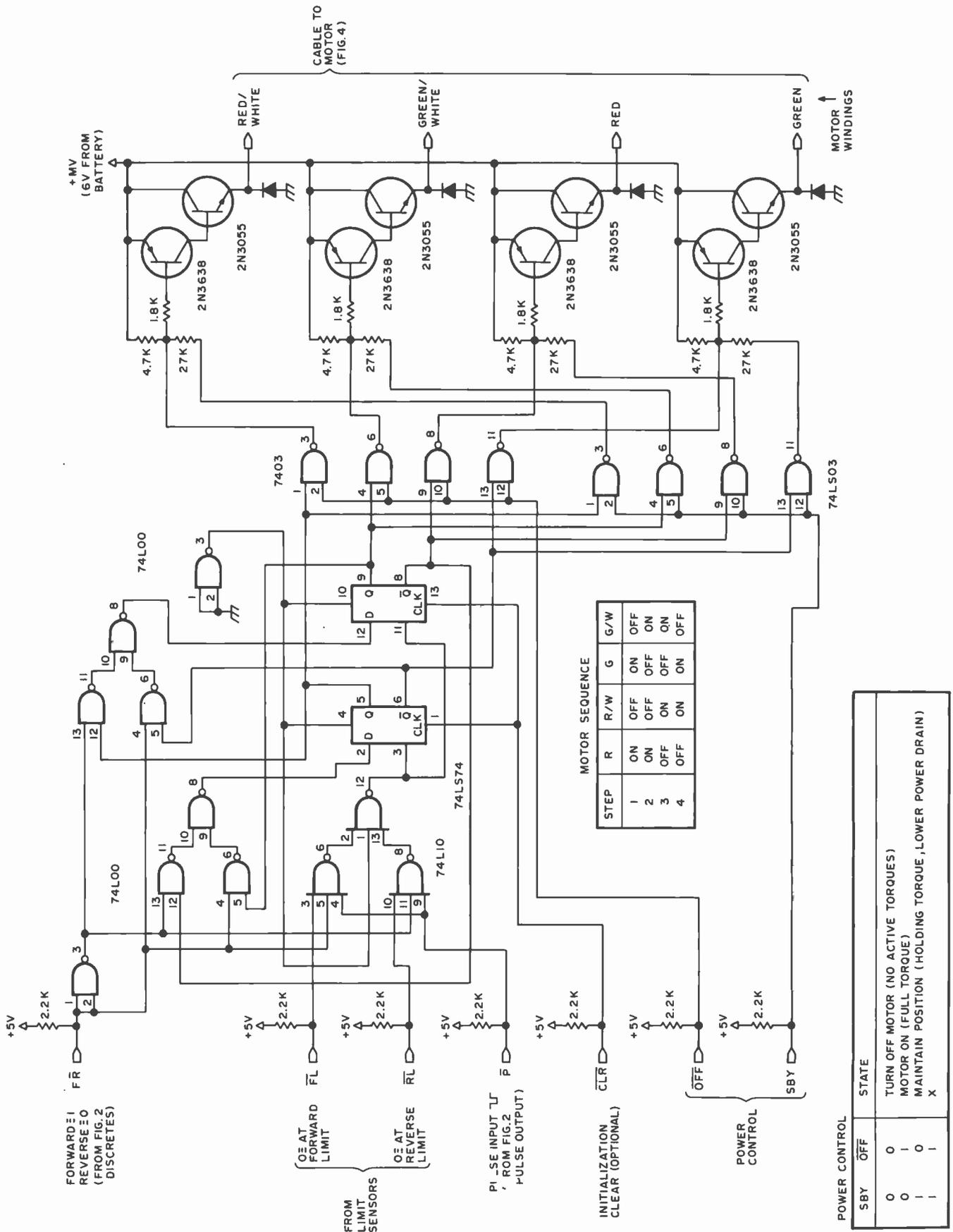
Connections to a Slo-Syn™ stepping motor are shown in figure 4. In the robot, all motors are powered directly from the main 6 V battery (+MV). Values for the resistance R can vary from zero to several ohms and are selected to be compatible with the motor's maximum current rating and dynamic characteristics.

Robot Control: Software Aspects

The previous discussion dealt with several general and specific aspects of the robot system and how it is intended to operate, once completed. Of the three major subsystems, motive, manipulator and sensory turret, only the motive subsystem is fully operational at present. The others are in various stages of detailed design and construction. Photo 4 reflects the current state of construction, which includes several temporary items such as the hand wired computer backplane and stepping motor drive modules, the crude "hand" made from sheet metal, and the cables leading to conventional power supplies. The 16 kg (35 pound) battery is on board to maintain balance, but not used, awaiting completion of the power conversion electronics. The open area at the front of the robot will be taken up by the manipulator assembly when it is completed. The power conversion modules and the sensory turret will go in place above the computer.

Software for the robot is created on the CDC-6400 computer using an 8080 cross assembler written in FORTRAN IV by Robert Mitchell. Object code for the 8080 is punched on 8 level paper tape by the 6400

Figure 3: Each stepper motor (there are seven presently incorporated into Newt's design) requires a drive circuit as shown here. This circuit contains counting logic and high current drivers needed to create the 4 step sequence of signals required by a



Slow-Syn™ stepping motor. The inputs to this interface circuit include one level (\overline{FR}) signal and one pulse signal (\overline{P}) derived from an output port such as the one shown in figure 2. The forward and reverse limit inputs are used to override commands when a sensor detects that rotation of the motor has reached some mechanical limit of the motor's activity. The power control logic (summarized in the chart) is used to conserve battery power by allowing a "standby" mode for the motor when it is not being actively driven. This standby mode holds the motor shaft position actively, but uses less current than the full torque state used to move the shaft from step to step.

system, and is subsequently read into the robot's memory through an ASR33 Teletype. A modest resident monitor program is stored in the first 666 bytes of EROM, which allows dumping and modifying of memory, punching and reading of paper tape, and branching to any memory location. The monitor is essential, because the computer has no conventional "front panel" with switches and lights.

The current robot control program occupies about 1000 bytes of programmable memory, not including table areas of variable size. It provides a method for exercising direct control of the robot for testing purposes. The coding, written by Dennis Toms, is very general, modular and compact. In order to start and stop the stepping motors without error, it is necessary to provide a profile of acceleration and deceleration. This is most conveniently done by using a time-delay table giving the appropriate time intervals between stepping motor pulses. The table is precomputed to yield a uniform acceleration over some time period. It is sufficient to have a single table for all motors. Each motor has associated with it a 14 byte motor status word (MSW) whose format is given in figure 5. The motor number and motor flag each occupy one byte; the other entries are two bytes (one word) each. The motor number byte is a fixed number which associates the MSW with a particular motor. The motor flag byte is a code which gives the current state of the motor (off, on, standby, accelerating, decelerating, emergency stop, etc). The speed pointer word is an index which points to the current entry in the time delay table. The top speed word specifies the index to the delay table entry, giving the shortest permissible delay between stepping motor pulses (maximum motor speed). The total steps word specifies the number of steps the motor is to execute. The acceleration, constant speed and deceleration counters specify the number of steps to be taken in the acceleration, constant speed and deceleration phases of the motion, respectively. These counters are decremented appropriately as the motion takes place, so that if necessary, at any time during the motion the state of the motor can be determined by

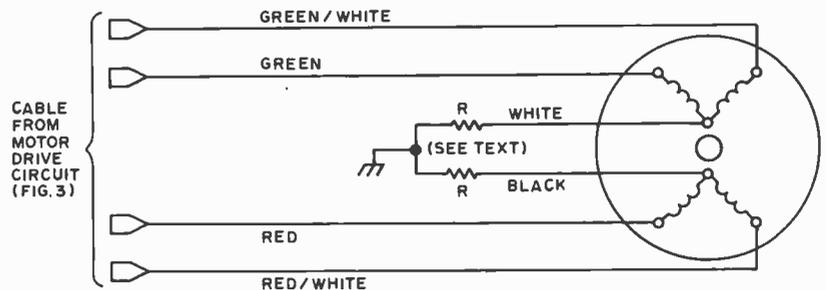


Figure 4: Wiring and color code for the Superior Electric Slow-Syn™ bifilar stepping motors. The resistance R should be set (see text) to reflect typical operating characteristics of the motor's use. The wiring is shown by color designations, and is connected to the drive circuit of figure 3. The ground return to figure 3 can be via the robot's chassis or through an additional circuit in the cabling.

interrupt handling routines. After all bytes of the appropriate MSW have been loaded with proper values, low level software takes over to carry out the motion. A pulse is issued to the selected motor causing it to advance 1.8 degrees. After a delay specified by the first entry in the delay table, a second pulse is issued, and the speed pointer is incremented to point to the second value in the delay table, and the acceleration counter is decremented. This process continues until the acceleration counter reaches zero. When this occurs, the acceleration phase is complete and the constant speed phase is entered. For this phase, the speed pointer is

Figure 5: Format of the motor status word. There is one motor status word allocated to each stepper motor; an interrupt driven process updates and times the operations of the motors. The information includes motor identification, status flags and parameters which specify the details of a cycle of operation consisting of acceleration, constant running speed and deceleration.

Motor Status Word (MSW)	
Motor Number	Motor Flag
Speed Pointer	
Top Speed	
Total Steps	
Acceleration Counter	
Constant Speed Counter	
Deceleration Counter	

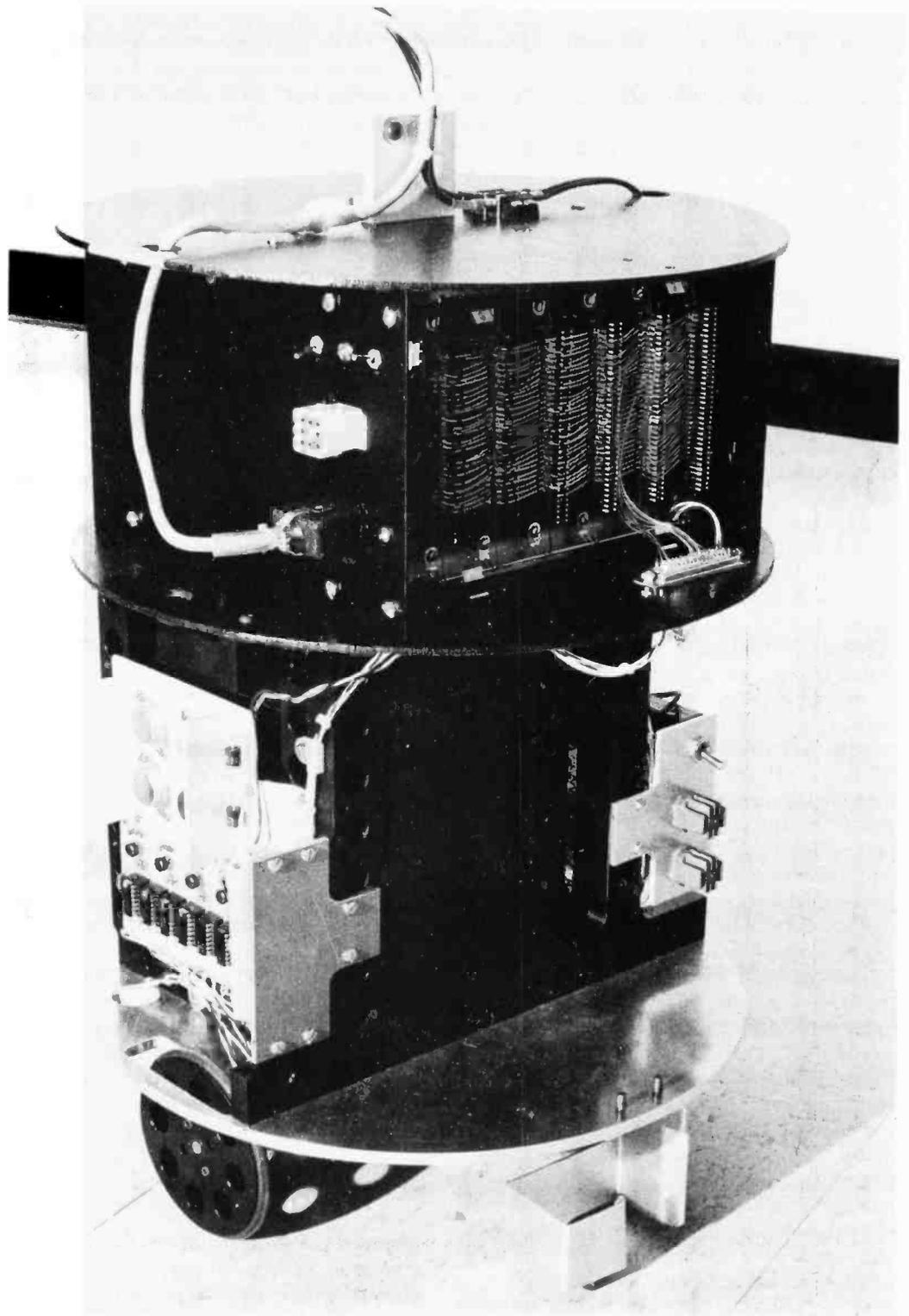


Photo 4: The robot Newt as it appears in the current state of construction. The hand wired computer backplane and stepping motor drive modules and sheet metal "hand" and power cables to an external source are all temporary.

held fixed, pointing to the time delay value corresponding to the top speed index, while the constant speed counter is decremented for each step. After the constant speed phase is completed, the speed pointer is decremented for each pulse, causing the motor to decelerate. The motor finally stops after a number of pulses have been issued equaling total steps. The motor is then placed in

standby condition and this fact is stored in the motor flag byte. If unforeseen conditions arise which make it inadvisable or impossible to complete the motion as planned, the interrupt software stores the appropriate emergency stop code in the motor flag and the motor is halted. In all but extreme emergency cases requiring instant stopping of a motor running at high speed,

the software can recover the total number of steps actually taken, saving it for navigational purposes.

To exercise direct control of the robot, a simple interpretive command system was developed. The currently implemented commands are listed in table 1. Each command consists of the 1 byte ASCII character F, B, L, R, W, V or Q, followed by a 2 byte argument. In the case of the F, B, L and R commands, the argument is simply the number of motor steps to be taken (less than or equal to $2^{15} - 1$). For the W command, the argument is the waiting time in units of 10ths of a second. For the V command, the argument specifies the new maximum speed of the robot in steps per second. For the Q command, the argument gives the 16 bit address of a sequence of commands stored in memory. Of course, many other commands, such as those appropriate for sensory turret and manipulator motion, will be added to the list of table 1. A program of robot activity consists of a simple sequence of these 3 byte commands, one after the other. For example, the following 15 byte sequence (arguments shown in hexadecimal notation):

```
F 0A10
L 000B
B 0010
W 0066
B 0002
```

would cause the robot to move forward 2576 steps (about 1.4 meters), turn left 11 steps (a few degrees), back up for 16 steps, stop and wait for 10.2 seconds, and then back up two steps. All accelerations and decelerations are taken care of automatically by the lower level software.

Commands can be given to the robot in several different ways (command modes) which are specified in table 2. For the D mode, the robot simply executes the commands directly as they are entered on the Teletype. After each command is given, it is necessary to wait until it is carried out before entering another command. In the R mode nothing is executed, but each command is recorded in sequence in the memory. The N mode is a combination of the D and R modes. It permits a sequence of commands to be recorded as they are being executed. The P mode allows playback of any previously recorded sequence. In this mode, the Teletype cable can be disconnected to allow the robot to roam freely. If the C mode is specified, the robot creeps along at a very slow pace under an F, B, L or R command for an indefinite number of steps. The motion is terminated by de-

Table 1: Currently implemented commands. The demonstrations of operation seen in photo 5 were created using an interpretive sequence including these seven basic commands. The command list is open ended in that many additions to the possible operations are expected as the software is developed further.

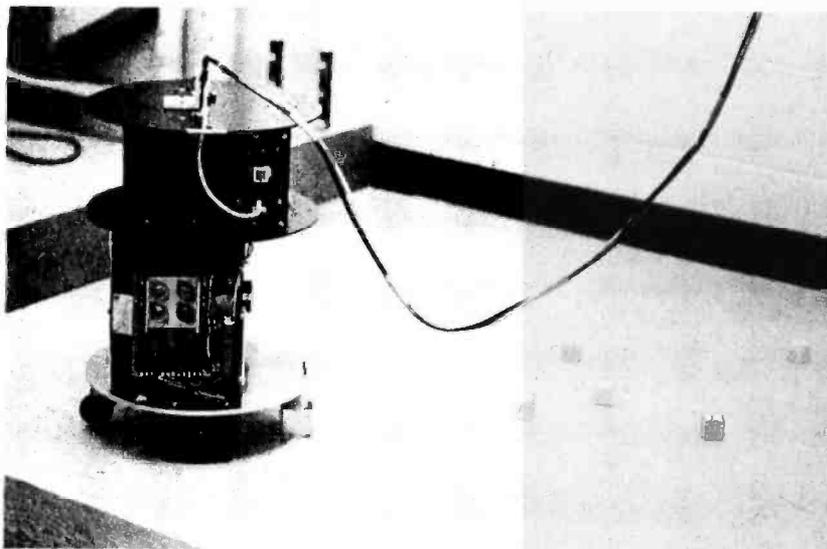
ASCII Command Code	Command Definition
F	Move robot forward
B	Move robot backward
L	Rotate robot left
R	Rotate robot right
W	Make robot wait
V	Set robot maximum speed
Q	Execute command sequence

pressing any key on the Teletype, after which the number of steps taken is printed out. The T mode is a combination of the C and R modes. The E, L and O modes are "bookkeeping" in nature, and permit erasing, listing, and setting the origin of a command sequence, respectively. During the creation of a program of action for the robot (command sequence), any combination of modes can be used as desired. It is easy to think of many other useful modes which could be added to the list of table 2.

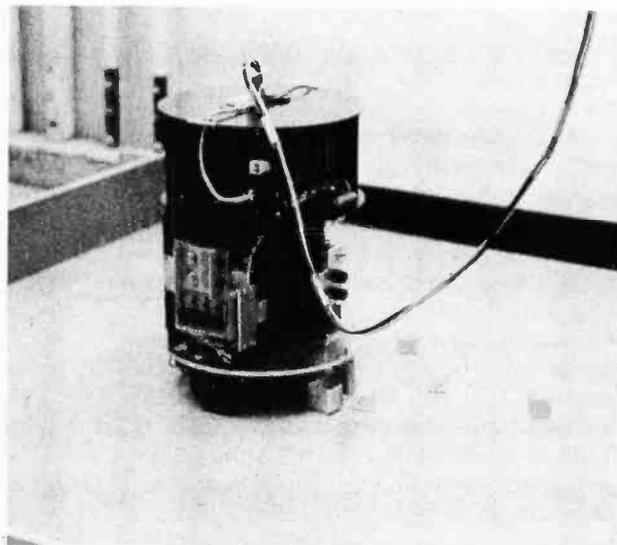
Using the commands listed in table 1, it has been possible to do a number of preliminary experiments with the robot which serve to test both the hardware and software, and to prove some of the fundamental ideas. For example, one can place wooden alphabet blocks on the floor at random, and then program the robot (using the N and T modes) to "pick them up" by trapping them in its temporary "hand." After putting the blocks and the robot back in their initial positions, the playback (P) mode can be used to repeat the motions automatically. One such test, rather in the form of a demonstration, is shown in photos 5a to 5g. In the first frame, photo 5a, the robot is moving towards the nearest block from its

Table 2: Command modes for the robot. The software presently implemented is used to try out various exercises of the robot's machinery. The software is structured into several command modes described by this list.

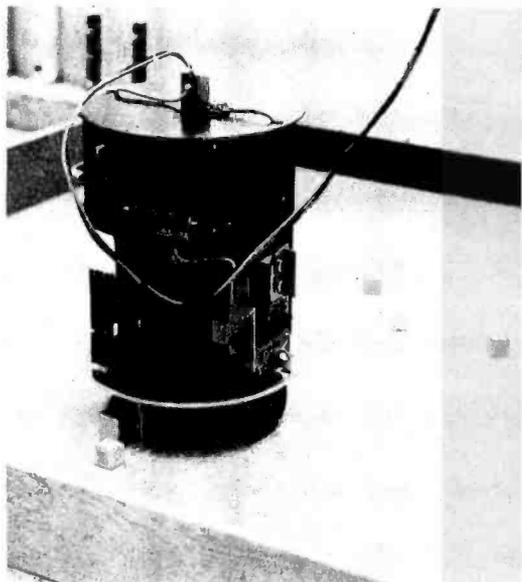
Command Mode	Type of Operation
D	Direct (simply execute the commands)
R	Remember (record commands)
N	Normal (combine D and R)
P	Playback (run through a command sequence)
C	Creep (directly execute motions at a turtle's pace)
T	Teach (combine C and R)
E	Erase
L	List
O	Set origin



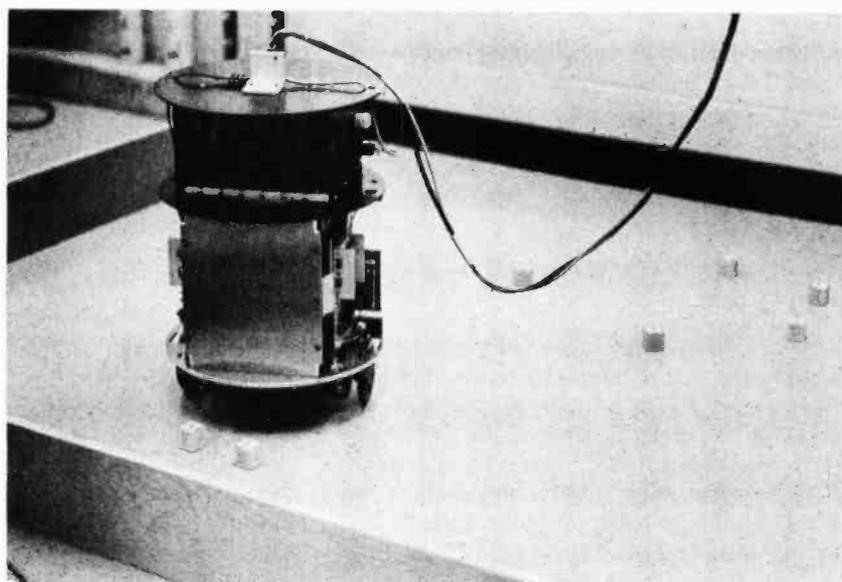
(a)



(b)



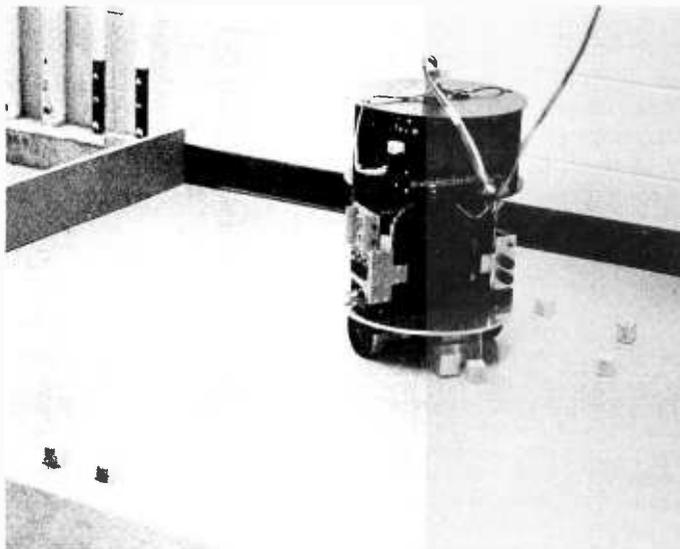
(c)



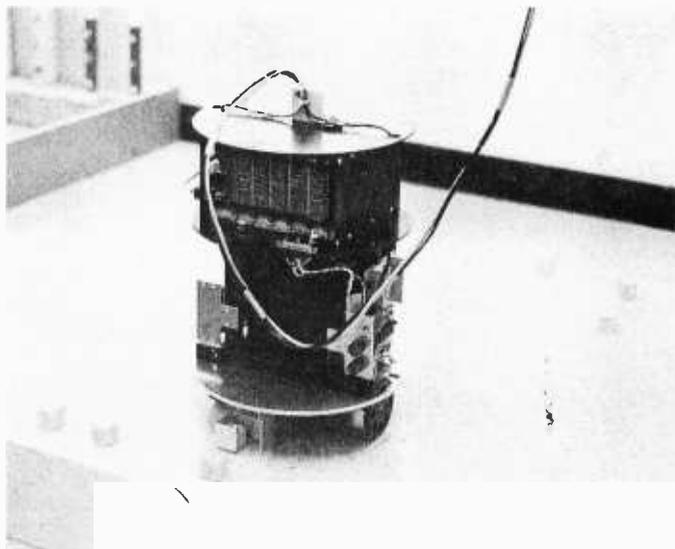
(d)

Photo 5: A program of action executed by the robot. In this sequence, prearranged blocks are pushed to appropriate locations to spell out N E W T, which is the robot's name.

- (a) The robot's actions start with its movement from an initial position in the left corner of the playpen toward its first block.*
- (b) As it closes in on the block, it maneuvers into position to entrap the block. Note that in this prearranged course, the robot is executing a fixed sequence of interpretive commands, and is not yet using visual inputs to find blocks at arbitrary positions.*
- (c) After trapping the block and moving to the final position, Newt releases the block at its final position.*
- (d) Here, a second block has been fetched and released next to the first block, and Newt is turning around in order to head for the third block.*
- (e) Now Newt is ready to pick up the third block in the sequence.*
- (f) After placing the third block and going to the fourth, Newt has picked up this last block.*
- (g) Finally, the fourth block has been placed, completing the spelling of Newt's name, which is the end of the programmed sequence of approximately 100 interpretive motion commands.*



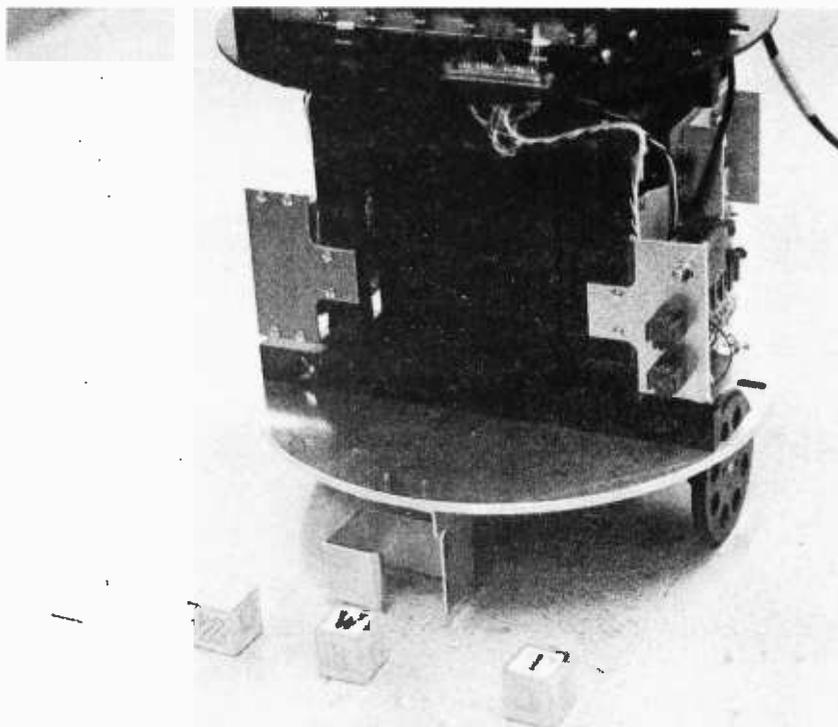
(e)



(f)

initial position in the left corner of the playpen. In 5b, it is maneuvering into position for "picking up" the block. In 5c, the block has been picked up, moved and released in a new position. In 5d, a second block has been fetched and placed next to the first block, and the robot is turning around and heading for a third block. In 5e, it is about ready to pick up the third block. In 5f, the third block has been placed, and the fourth block is in the "hand." In the final frame 5g, the fourth block has been placed, spelling out N E W T, which is, of course, the robot's name. Approximately 100 3 byte commands were required for this sequence. The program can be executed dozens of times without error, and the final positions of the blocks and the robot after the sequence is finished has a scatter of a few millimeters ($\pm 1/8$ inch) in each direction. Of course, when the robot is complete with all its senses, it will seldom have to execute such a lengthy sequence in a completely open loop fashion. Perhaps it can eventually do tasks such as searching for and finding specific alphabet blocks in a random pile of blocks. This would require frequent closing of the feedback loops through the visual senses.

As the robot goes through its motions, such as depicted in photos 5a to 5g, it seems to possess an almost uncanny grace and precision. Small children, when watching it, are frightened at first, but this soon gives way to playful interest and warm curiosity. Even hardened computer experts are amazed to see a computer driving itself around on wheels!



(g)

Some Personal Remarks on Building Robots

It will not be possible for one person alone to write the software for the robot. This is far too large a job. It is hoped, however, that as the hardware nears completion, it will act as a focal point for many persons wanting to experiment with the robot by writing their own software. A

fascinating project would be to create a general "Robot Control Language" which would free each programmer from the details of the hardware. What a rich experience it would be to work together, exchanging ideas in a highly interactive way. There are many tutorial possibilities for computer

science classes and beginning students learning about computers. There are problems in psychology, procedural languages, human-to-machine communication, functioning of parts of the brain . . . (Working with robots is certainly one way to gain a much greater appreciation for the complex-

SOME TERMINOLOGY

Amp hour: A unit of energy for rating batteries. The battery of NEWT, with its 84 Amp hour capacity, can store enough charge to drive a steady 1 A load for 84 hours, or a 14 A load for six hours. The finite capacity of any practical battery means that any mobile robot must incorporate some programs for seeking electrical outlets and recharging batteries periodically.

Azimuth: As used here, an angle relative to a fixed direction in the horizontal plane.

Cognition: As Webster has it, this is "the act or process of knowing, including both awareness and judgement." [*Webster's New Collegiate Dictionary*, 1976 edition.] In the context of robots and artificial intelligence, this term refers to programmed models which approximate the behavior of natural cognition.

Degrees of freedom: The state of a robot mechanism (or any other system) can be described by specifying the current value of each variable parameter. Thus, if a robot arm has seven joints, the position of its "hand" might be determined by the angular setting of each joint. Each such independently variable parameter of a system is called a "degree of freedom," so the seven jointed arm would have seven degrees of freedom.

Heuristic: A heuristic computer program is one which starts out with an approximate method of solving of a problem within the context of some goal, and uses feedback from the effects of the solution to improve upon its own performance. Heuristic programming is one of the major contemporary artificial intelligence techniques, and is a key to developing a cognitive robot.

Manipulator systems: A generic term for any mechanical device which a robot uses to directly manipulate its environment. In the NEWT robot, this is currently (see photos) a simple sheet metal frame which can catch a block and slide it across the floor as the robot moves, with no active grasping; NEWT is intended to eventually have a much more flexible system of manipulation as described in the text. Most industrial robots currently in use consist of manipulators alone, without much in the way of sensory feedback or motive systems.

Motive systems: A generic term for the mechanisms used to convey the robot around its environment. In the NEWT robot, this refers to the two drive wheels, balancing caster and stepper motors which propel the robot.

Open loop, closed loop: A closed loop system is one which operates with feedback from errors. The feedback is intended to correct for the errors and thus approach the truth; an open loop system ignores error signals and operates on the sometimes

naive assumption that no errors occur. The terms must be qualified by a reference to the time intervals involved in the system: NEWT, for example, is a closed loop system over long time periods, since it is intended to navigate using feedback from its sensors; however, due to the processing loads associated with sensors, NEWT operates open loop between navigation sightings in a manner analogous to the dead reckoning method of navigation used occasionally by airplane pilots or captains of ships.

Round off errors: In operations such as addition, multiplication or calculating transcendental functions, there is often some uncertainty in the least significant part of the result. In an extended calculation in which these operations are repeated over and over, appreciable round off errors can accumulate. In a digitally controlled vehicle guidance system such as that used for a robot, these numerical errors are a major source of uncertainty in the vehicle position, and are just as important as more obvious sources of error such as step quantization or slippage in the drive mechanism.

Senses: In a robotics context, senses are specialized peripherals which convert information about the environment into signals which can be analyzed by a computer or used directly by the electronics, as in a reflex. Sensory information may be obtained from devices as simple as a microswitch with a "feeler" arm, or as complicated as photoelectric imaging arrays with zoom lenses and pointing mechanisms.

Stepping motors: An ordinary electric motor is characterized by continuous motion when energized. A stepping motor uses a different design philosophy to achieve a motor which will move its shaft in small incremental angular steps on command, and will actively maintain its position in between each command. This type of motor is very well adapted to digital control of mechanical systems, and is used by NEWT for all mechanical motions in the robot.

Step quantization: The stepper motors have a finite angular resolution built into their design. This means that any mechanical motion derived from the motor will have a certain minimum step size, so that any attempt to position to a finer tolerance must be approximated.

Trajectory: The path of a moving object is its trajectory. In the case of the mobile robot, a trajectory is planned before motion takes place, given a desired goal position and a world model which covers its course and objects which may be in the way.

World model: A world model is the result of cognition as implemented in robots. Formally, it is an information structure built up in the memory of the robot, based on both initialization and heuristic interaction with the environment.

ity and capability of the brain.) In addition, experiments can continue on sensor development and interfaces from sensor to computer. There are a great many practical spin-offs from this kind of work.

Many people believe that as more and more advances are made in microelectronics, the prospects of mass producing robots will become attractive, and the prices of these hypothetical machines will plummet. (Let us hope we will have learned something from *RUR*.) If this occurs, many applications will open up. Besides such things as planetary surface exploration, such as already demonstrated by the Viking robots, one can envision undersea robots working on oil pipelines and well heads, coal mining robots, fire fighting robots, agricultural robots, robots on assembly lines producing customized articles, robot-like prosthetic devices, and many other types of robots for specialized and general service, doing jobs which are too difficult, too dangerous, or which are otherwise undesirable for humans.

I have been asked many times the questions, "Why are you building a robot?" and "What will it do when it is finished?" The answer to the second of these questions is easy: I simply don't know what the robot will be able to do. This is the whole point of building the robot. Given a modest amount of hardware and a greater amount of software, thoroughly integrated to form a system, the idea is to find out just what such a system is capable of doing. The whole is likely to be far greater than the sum of the parts. The system is pushed as far as the available time, money and energy will allow in order to learn what can be done and what cannot be done; in other words, to explore the frontier of robot research, and to know and understand the problems involved. The necessary knowledge can neither be obtained by theoretical studies, nor by simulations using large computers.

In regard to the first of the questions, I have been fascinated with robots since the mid-1950s and have constructed several robot devices prior to the one described in this article. The construction of such machines presents many interesting challenges. A functioning robot is a most curious blend of electronics, mechanics, computer design, computer programming and artificial intelligence. All these fields come together in the design and construction of a robot, and each must be explored in depth. Added to this are the challenges and excitement of locating obscure components in surplus store parts bins around the country, planning, building, and then replanning, rebuilding, and, of course, experimenting and learning. To me,

About NEWT's Name and Family Tree

The origin of NEWT's name is buried in an often quoted verse from *Macbeth* by William Shakespeare . . .

*In the cauldron boil and bake;
Eye of newt and toe of frog,
Wool of bat and tongue of dog,
Adder's fork and blind-worm's sting,
Lizard's leg and howlet's wing,
For a charm of powerful trouble,
Like a hell-broth boil and bubble.*

Newt I, the present robot's predecessor, was a light-seeking robot consisting of a large eye on a stalk rising above a motor driven platform.

all these are fascinating aspects of endeavors which are perhaps best left to amateurs. I say this, because I strongly believe that the amateur computer enthusiast has a golden opportunity to participate in advances in the field of robotics. In fact, the amateur has several advantages over the professional. The research can be as abstract as the amateur wishes it to be and can be conducted without regard to immediate payoff potential in the marketplace. There is no need to spend time writing elaborate proposals, no need to continually justify the direction of the work, and no need to get hard results every few months to write up and stick into quarterly reports. History has shown that precisely this atmosphere of freedom which surrounds the amateur is the atmosphere in which brilliant innovations and discoveries are sometimes made. ■

Acknowledgment: The author wishes to express gratitude to Dennis Toms for his enthusiastic help and interest in this project.

Photo 1: The complete setup of the IBM Selectric Keyboard Printer, typing under the control of a KIM-1 microcomputer with a 4 K memory expansion. The Selectric interface described in this article is housed in the equipment case in the center of this photo.

Interfacing the IBM

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Hamilton Hall C-23
Harvard Business School
Boston MA 02163

Photography by Carole Brock

One of the most desirable forms of computer output is high quality typewritten text suitable for preparing letters, reports and other documentation. A word processing system which speeds up the process of writing and revising text would be a very useful and feasible application for a small microprocessor based system, provided that a suitable hard copy output device can be found at a reasonable price.

An ideal output medium for such a word processing system would be an IBM Selectric office typewriter. Selectrics are moderately expensive when compared to ordinary typewriters (\$630 to \$830 depending on the options chosen), but they are ubiquitous in the office environment, produce very high quality typed output, and can be used to print in many different type styles simply by changing the ball shaped typing element. Special typeballs are available for printing mathematical symbols and for the APL character set (see "What is APL?", by Mark Arnold, November 1976 BYTE, page 20).

Unfortunately, the job of converting a Selectric office typewriter is made somewhat more difficult by the fact that (contrary to popular belief) the Selectric mechanism is entirely mechanical and not electronic in nature. The only use of electric power in an ordinary Selectric is for the motor which turns the drive shaft and various gears and cams. It is necessary to use solenoids to push levers and "bails" in the base of the mechanism to achieve printing under computer control. Similarly, contact switches must be installed in order to use the keyboard for computer input.

There is another alternative, however. A variety of computer terminals and other devices based on the Selectric mechanism are becoming available on the surplus market, often at a fraction of their original prices. These machines have their own built-in solenoids or other means for mechanical control, and present some sort of electrical or electronic interface to the outside world. The simplest, most commonly available, and of-



Selectric Keyboard Printer

(Teaching KIM to Type)

ten the cheapest of these are the Selectric Input/Output Keyboard Printers, Models 73, 731, 735 and others. They were manufactured by IBM, typically for use as IO devices in other companies' computer systems. As these systems have become obsolete, the Selectric Keyboard Printers have found their way into surplus channels.

As a business school student and experienced user of computers, I have always wanted to build a word processing system around my own home computer. Hence I seized a chance to acquire a Model 73 Keyboard Printer for \$450 from the Computer Warehouse Store in Boston. (These units were sold out in a few weeks; I have heard of prices ranging from \$250 to \$1500 through other channels, but as interest in the units increases, their typical prices are bound to rise.) Armed with a couple of old IBM manuals provided by the Computer Warehouse Store, I set out to accomplish what I expected would be a straightforward interfacing process.

This article is a report of my experience, and a detailed description of the interface which I built. Briefly, the interfacing process, while simple in principle, was not at all straightforward in practice. But it was successful, even for such a mechanically inept person and relative novice in electronics as me. For about \$50 in parts (including such extravagances as a pretty cabinet and a \$20 IBM connector to plug into the Selectric's peculiar 50 pin receptacle), and lots of labor, I produced the unit shown in photo 1. It's only an interface to the Selectric printer, since I'm content to use my existing ASCII keyboard for input. It has its limitations, but it works.

This, of course, is hardly the last word on Selectric Keyboard Printer conversion. As a BYTE reader, I would be delighted to see information on more comprehensive interface designs, as well as actual experiences with several of the units currently on the market. Since most of them are sold on an "as is" basis, these machines can bring

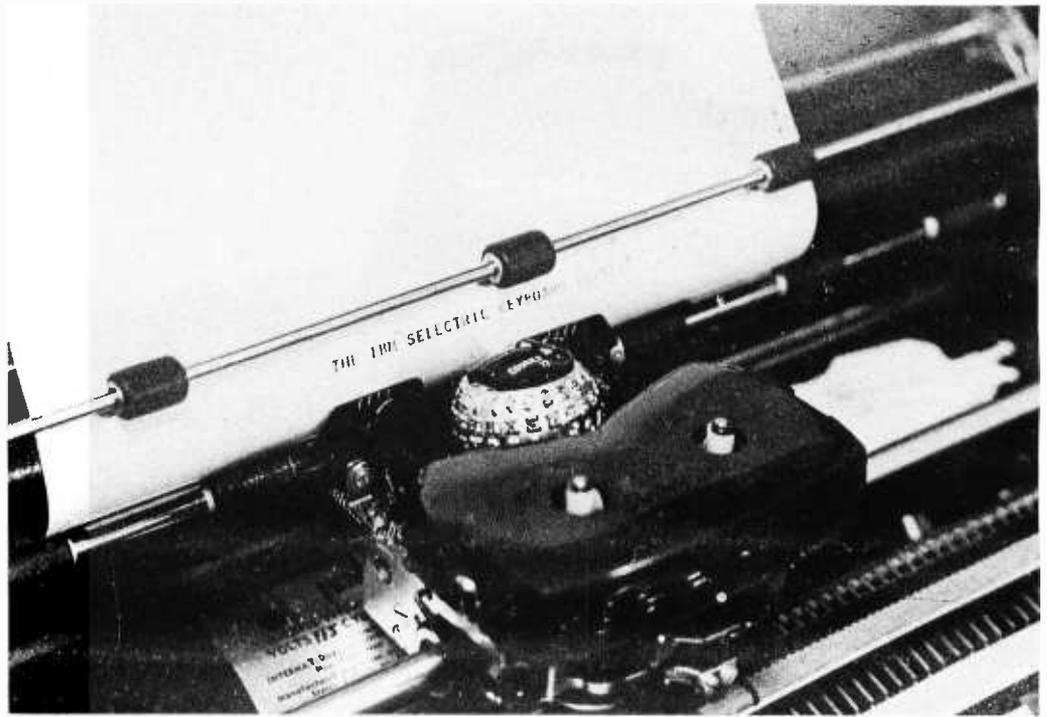


Photo 2: A closeup of the IBM Selectric ball mechanism on its moving carriage within the printer. The Courier 72 ball shown here is one of many balls available with the "Correspondence" coding arrangement.

you a lot of grief (read on). So it's wise to report on problems, and how you overcame them.

The Selectric Mechanism

To appreciate what the interface must do (and what can go wrong), it is first necessary to understand basically how the Selectric mechanism works. The typing element is a

golfball sized hollow sphere embossed with up to 88 characters, arranged in four horizontal rings of 22 characters each. Photo 2 illustrates the ball in its rest position in the mechanism. All the lower case or unshifted characters appear on the "western hemisphere," the side normally closest to the paper. The upper case characters are in corresponding positions on the back side or "eastern hemisphere." Pressing the shift key causes the whole typeball to rotate 180°, thereby allowing the upper case characters to be printed. Hence, the actual typing operation can select any of 44 characters, four half rings of 11 characters each, with five to the left and five to the right of the center or "home" position on each ring. A particular character is selected by causing the typeball to tilt up or down and rotate right or left; then the ball jumps forward to strike the ribbon and paper. These movements account for the peculiar "dancing" motion seen when the Selectric is typing continuously. The typeball is mounted on a carriage which moves across the page, as opposed to traditional pre-IBM typewriters where the paper carriage moves and the typing mechanism remains stationary.

The actual tilting and rotation of the typeball is accomplished by an incredibly complicated system of latches, pulleys and levers which are driven by six moving "bails," or rods in the base of the machine. Although we need not understand the detailed mechanical linkages, we should appreciate the roles played by these six

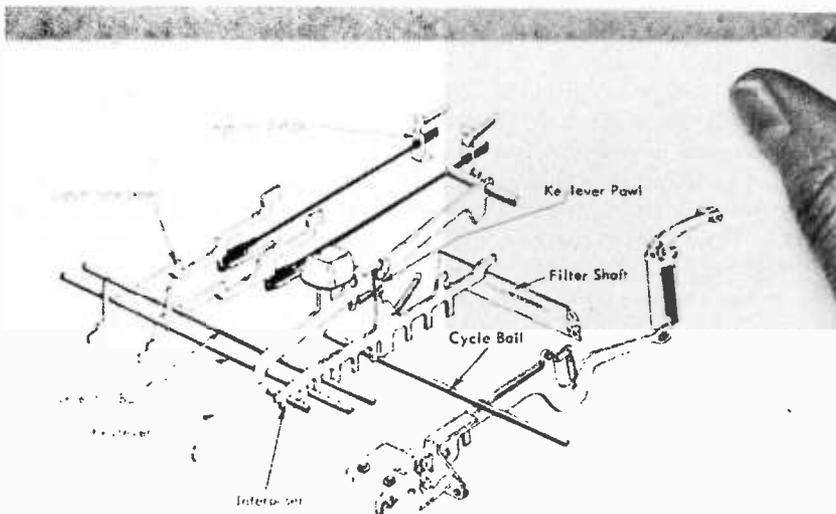


FIGURE 35. Keyboard Section and Character Selection

Photo 3: This diagram, from the IBM manual number 241-5159-3, shows how the various bails of the mechanism are connected in a typical case. The IBM manuals for the typewriter output unit are valuable reference materials and can be obtained by calling your local IBM office.

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OSI'S DREAM MACHINE is a totally integrated computer system.

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In the configuration shown above, the Challenger includes everything an end user needs for a complete small computer system. All you add is 110 VAC power and a desk to put it on.

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OSI Challenger 65 with 16K RAM, serial interface, system monitor PROM, and floppy disk bootstrap PROM.

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The Challenger system is complete, fully assembled and configured so that the Disk Operating System can be booted in immediately on system power-up. Even a relatively inexperienced operator can have a complex BASIC program on-line just seconds after the system is turned on. The ease of use, high reliability, and large library of standard BASIC applications programs make the OSI Challenger System the first practical and affordable small computer system for small business, educational institutions, labs, and the personal computerist.

PRICES:

Challenger System, complete as stated above with terminal and monitor

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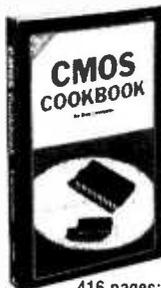
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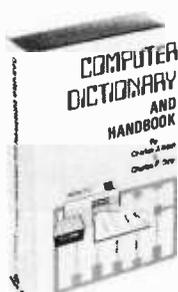


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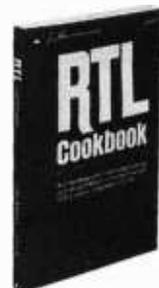
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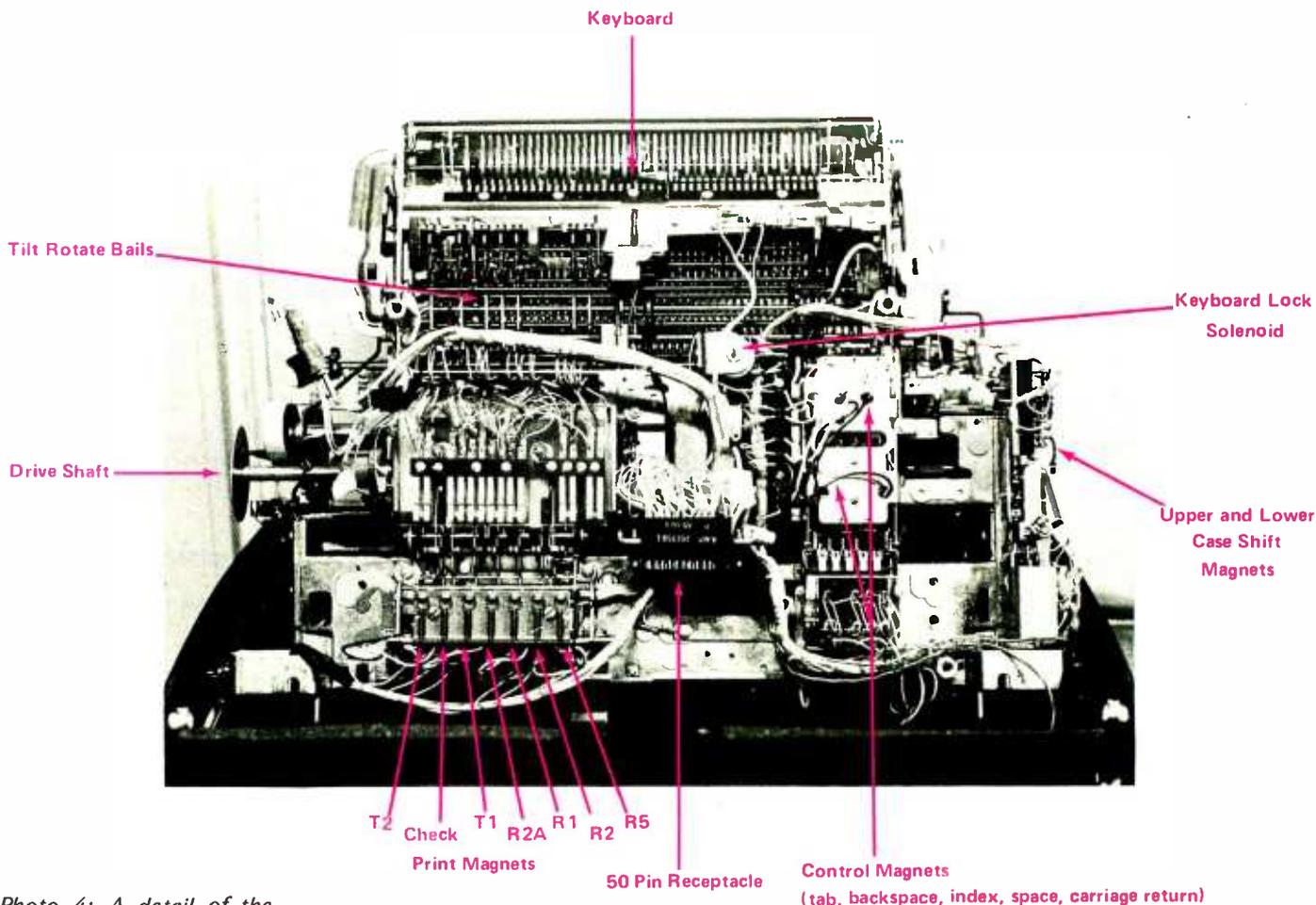


Photo 4: A detail of the underside of the Selectric Keyboard Printer with housings removed. The overlay shows several of the key points such as the location of various magnets, the switch contacts and interconnection receptacle.

must actuate the “cycle bail” to start the printing process. Hence a trip mechanism is provided which moves the cycle bail whenever any of the armatures is pulled down. However, there is one character on each hemisphere which should be printed when none of the magnets is energized, for the code 000000. Hence the trip mechanism is connected to a seventh magnet, called “check,” which provides an odd parity function for the other six magnets. It is energized whenever necessary to ensure that the total number of magnets energized is odd. Thus the check magnet is energized on the code 000000, and this serves to actuate the cycle bail. (I didn’t realize this when building my interface, so I can’t print those two characters yet. Don’t make the same mistake!)

Besides the print magnets, there are a number of other magnets and armatures inside the Keyboard Printer which control special functions such as space, backspace, tab, carriage return, index (ie: advance paper without returning), ribbon shift, and upper and lower case shift. Many of these magnets

can be seen in photo 4, which exposes the underside of the machine and outlines the positions of many components. The upper and lower case shift magnets are latching, and hence they lock the machine into the new case until the opposite magnet is energized. Note that the operator cannot shift the machine back into lower case when the upper case magnet is latched! By Murphy’s Law this is bound to happen whenever you are testing the interface, but it can be remedied by fooling around with the shift cam at the end of the drive shaft.

No electric power is provided for any of these magnets inside the Keyboard Printer, but the coil connections are brought out to the 50 pin receptacle at the back of the machine. The magnets are rated for 43 to 53 VDC at 125 to 300 mA, applied for at least 10 ms in order to pull down the armatures and cause the desired action.

Switch Contacts

The other major addition to the basic

Continued on page 133



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Interfacing With an Analog World—Part 2

Last month we discussed transducers and amplifiers. These are necessary portions of a signal processing system which result in scaled voltages of, for example, 0 to 10 V corresponding to the original physical parameter being measured. But how can we convert these voltages into numbers inside a computer for computation, and use numbers from computations to control external voltages? In this article we'll see how some of the more common conversions are accomplished. We'll start with digital to analog conversion, even though this may seem at first glance to be backwards. The reason for starting with the output process is that digital to analog conversion is simplest, and that many analog to digital input conversion techniques require a digital to analog conversion as part of the process.

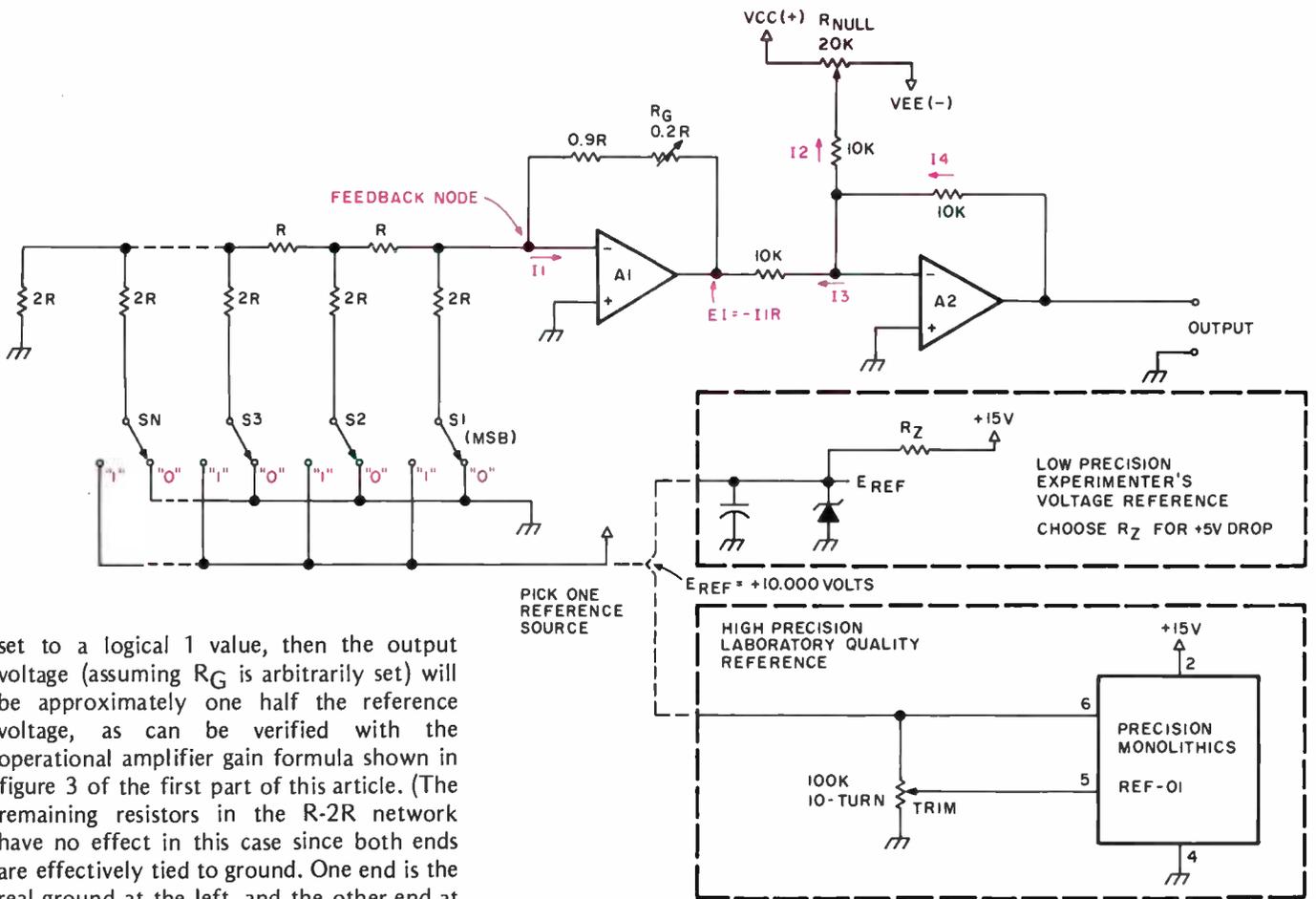
There are several techniques for making a digital to analog converter (often abbreviated DAC). In most cases these converters use some form of binary weighted current or voltage summation that is controlled by the digital word presented to its inputs. A typical example is the classical R-2R technique which is outlined in figure 1. The active element A1 is an operational amplifier of unity gain inverting follower configuration. Although an amplifier of the 741 general purpose family will suffice in many experimental situations, it is often better to select one of the more costly premium grade integrated circuit operational amplifiers. We have shown the digital inputs of the circuit in figure 1 as mechanical switches, a configuration which is most useful in a tutorial situation when teaching the concept of the digital to analog conversion. In practical digital to analog conversion applications the switches are electronic, and are

controlled by some form of n bit binary data source such as a counter or computer output port.

A precision reference voltage source is required as well, and for many commercial and industrial system designs this will be a precise 10.000...V. The accuracy of the converter is largely dependent upon the precision of the reference voltage. Although any precise voltage regulator circuit may be used for the reference, I have found in laboratory instrumentation experience that commercial products such as the REF-01C) or REF-01HJ (for severe applications) work quite well. In a homebrew experimenter's situation, where relaxation of tolerances is quite normal, a simple zener diode reference circuit will often prove quite adequate.

Now let's consider the circuit in figure 1 in more detail. What happens when various combinations are presented to the digital input? Suppose that all the input bits are in the low state, which means that they are connected to ground by the electronic or mechanical switches shown. The value of the output voltage is given by the product $I_1 R$. When all bits are grounded through the switches, the input current to the amplifier is zero, as can be deduced by tracing, noting that there are no nonzero inputs to the amplifier's feedback node. (In practical circuits, though, there may be some output voltage under these circumstances due to offsets in the operational amplifier itself. These undesirable offsets may be nulled out through an offset adjustment potentiometer arrangement, R_{NULL} in this circuit. See some of the tutorial design books on operational amplifiers for further elaboration of this detail.)

If the most significant bit of the word is



set to a logical 1 value, then the output voltage (assuming R_G is arbitrarily set) will be approximately one half the reference voltage, as can be verified with the operational amplifier gain formula shown in figure 3 of the first part of this article. (The remaining resistors in the R-2R network have no effect in this case since both ends are effectively tied to ground. One end is the real ground at the left, and the other end at the feedback node of the amplifier is its "virtual ground" for the signal.) The analysis of the next most significant bit and the remaining bits of the digital word is a bit more complicated, but the result is what might be expected. The bit controlled by S2 in figure 1 will contribute one fourth of the reference voltage to the output of A1; the bit controlled by S3 will contribute one eighth of the reference voltage. And for switch, or bit, n (where n starts at 1) the contribution will be $E_{REF}/(2^n)$.

Let us assume that we have an 8 bit digital to analog converter of the type shown in figure 1. The word at the input terminals is 11001011 and the reference voltage is precisely +10.000 VDC. What is the output voltage? The following calculation, which is easily generalized, shows how the value is derived:

$$E_o = 10 \times \left(\frac{1}{2^1} + \frac{1}{2^2} + \frac{0}{2^3} + \frac{0}{2^4} + \frac{1}{2^5} + \frac{0}{2^6} + \frac{1}{2^7} + \frac{1}{2^8} \right)$$

↑
Output

↑
 E_{REF}

Fraction Based on a Digital Word

$$= (10/2) + (10/4) + (10/32) + (10/128) + (10/256) =$$

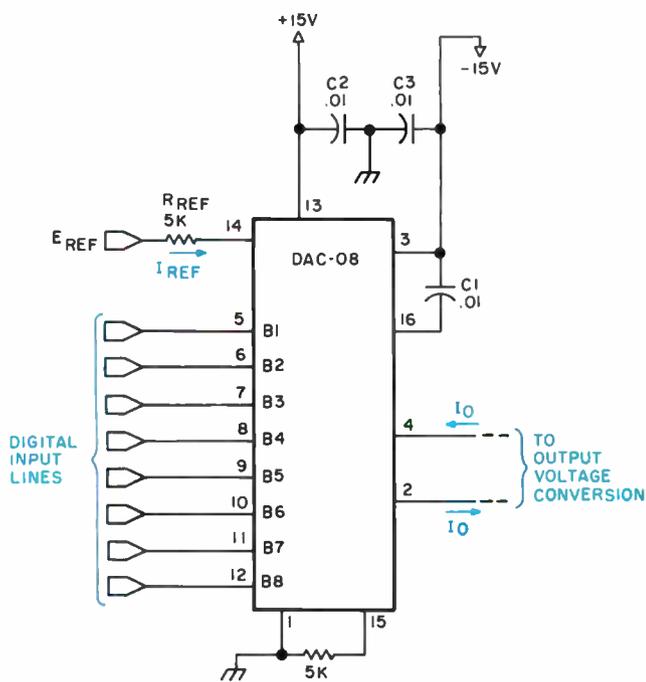
$$= 5 + 2.5 + 0.3125 + 0.078125 + 0.0390625 =$$

$$\approx 7.93 \text{ V}$$

But use of amplifiers and resistors as shown in figure 1 is hardly optimal in an age of integrated circuits. A number of manufacturers offer convenient low cost 8 bit integrated circuit digital to analog converters that contain almost all of the electronics, except possibly the E_{REF} supply and the operational amplifier used for output voltage conversion and level shifting. I have used those by DATEL, Analog Devices and Precision Monolithics with good results in laboratory instrumentation. Experimenters and designers will also find the parts in the Motorola MC1408 family, as well as several similar parts made by Signetics, to be quite useful. For my examples in this article I have selected the Precision Monolithics DAC-08. I found this product easy to obtain in low quantities (ie: one) through local distributors.

Figure 2 shows the basic circuit for using the DAC-08, along with two voltage conversion schemes for its current outputs of pins 2 and 4. The integrated circuit itself contains the electronic switches, the resistance ladder, a reference amplifier and the current output buffer that drives pins 2 and 4. Two types of input are required to make

Figure 1: A classical R-2R network digital to analog converter implemented as a circuit diagram with discrete parts and operational amplifiers. The essentials of any digital to analog converter are present: a reference voltage (two alternatives shown), a switched network that creates a binary weighted current controlled by the switches, and buffering and conversion amplifiers to create a voltage output which can drive other circuits.



OUTPUT VOLTAGE CONVERSIONS					
CONDITION	BINARY CODE	E _{OUT}	CONDITION	BINARY CODE	E _{OUT}
FULL SCALE	11111111	-9.96	(+)FULL SCALE	11111111	+9.96
HALF SCALE	10000000	-5.00	(+)ZERO	10000000	+0.04
ZERO	00000000	0.00	(-)ZERO	01111111	-0.04
			(-)FULL SCALE	00000000	-9.96

NEGATIVE OUTPUT
4B

BIPOLAR OUTPUT
4C

Figure 2: While the circuit of figure 1 could be constructed, it is usually more convenient to replace it with a monolithic integrated circuit device such as the DAC-08 part by Precision Monolithics, which is used for this illustration. The integrated circuit contains all the components of the R-2R network of switches and buffering amplifiers. It still requires an output conversion circuit (two variations shown) and a reference voltage. Because the R-2R network is an integrated circuit, this form guarantees the highest possible accuracy with none of the extra bother of hand wiring a circuit as in figure 1.

this digital to analog converter do its work. One is the reference current, I_{REF} thru pin 14. This current can be generated by a precision voltage reference (see figure 1) and a precision low temperature coefficient resistor in the ideal cases. (For low precision applications, ordinary resistors will work just fine.) For the configuration shown in figure 2 and a reference potential of 10 V, this resistor should be precisely 5000 ohms, a number which is derived from the documentation of the DAC-08.

The second major type of input to the converter is the 8 bit digital word that is applied through pins 5 to 12. In the notation of figure 2 and throughout this article, the bits are numbered from 1 (most significant) to 8 (least significant). Bit 1 of the DAC-08 package is wired to pin 5, with bits 2 thru 8 wired to pins 6 thru 12. The logic levels at the inputs are the usual TTL levels, with a low voltage (approximately 0 volts) signifying a logical 0, and a high voltage signifying a logical 1.

One of the output conversion circuits shown in figure 2 is a simple unipolar conversion which uses two resistors and no operational amplifier. With this conversion, the formulas discussed in figure 1 apply except for the fact that the voltage is negative with respect to ground. When the input word is binary 00000000, then the output of the converter system is 0.00 V. Half scale (-5 V output) is given by an input word of binary 10000000, and full scale output occurs when the input word is binary

11111111. The output under full scale conditions will not be 10 V, but approximately 9.96 V. This slightly unexpected condition is due to the mathematics of the switching network. Evaluating the formula for the conversion given earlier, with a reference of 10 V and a binary 11111111 digital value, we find:

$$E_{out} = 10 \times (255/256) = 9.96 \text{ V}$$

Also shown in figure 2 is an output circuit which uses an operational amplifier as a level shifter and voltage conversion device. Wired with the components shown, this amplifier connected to the DAC-08 gives a gain of 2 and shifts the generation of output levels to a symmetric bipolar range of -9.96 to +9.96 V. Note that it is impossible to get an exactly zero voltage in this case, since the 256 possible states are split symmetrically about zero. If the level shifting reference resistor R_{LS} were adjusted slightly off the 5 K value, the voltage range of the conversion could be pulled slightly (ie: 0.04 V) positive or negative so that a true zero would be possible for one of the binary states. There are other possibilities for the output conversion circuits and as in any design situation, a little bit of imagination always comes in handy. [Readers looking for more examples of typical applications should consult the applications notes of the various manufacturers. Of particular use is the excellent specification sheet and application notes on the MC1408 DAC, published by Motorola . . . CH]

Analog to Digital Conversions

With the concept of a digital to analog conversion covered, it is now possible to consider the opposite case: conversion of measured voltages from the sensor preamplifiers into numbers which can be processed and used by a computer. Of the many techniques which are available for performing analog to digital conversions, we will only consider the details of integration, counter (or ramp), and successive approximation methods here. These are the simplest and most universal methods.

One of the basic parameters to be considered when talking about any analog to digital conversions is speed. This is not a major consideration in the output problem already discussed, since digital to analog current output conversions essentially take place at the switching speeds of digital logic, and are then limited only by the final operational voltage output amplifier's response. In the input case, however, some form of approximation cycle which converges upon the digital value is required; as a result, the conversion can be somewhat slower.

Integration Methods

At the slow end of the analog input conversion spectrum is the integration method. This is the type of conversion which is typically used in digital panel meters and similar instrumentation. These

can be useful in cases where you might mount the digital panel meter or multimeter in a system, both as a readout mechanism and as a measurement conversion device. Many such instruments offer parallel digital outputs on their rear panels, along with control and strobe lines. The appeal of this approach often is affected by two characteristics: relatively slow conversion speeds and binary coded decimal (BCD) encoding directly taken from the displays. The relatively slow conversion rates become a problem when looking at signals other than "slowly varying DC levels" of very low frequency sources. The coding characteristics may in fact be optimal for many computational schemes in a computer program, but it can be a nuisance if one attempts to use such a meter in a binary oriented hardware system. The typical "dual slope" integrator used in these digital panel meter circuits is illustrated in figure 3.

The dual slope conversion circuit consists of five basic sections: an integrator, a comparator, a control logic section, a binary counter, and a reference current or voltage source. The integrator consists of an operational amplifier connected with a capacitor in a negative feedback loop. This capacitor is charged by the operational amplifier output voltage. The input to the integrator is taken from either the analog input or the reference source. The comparator is made with an operational amplifier that has an open feedback loop.

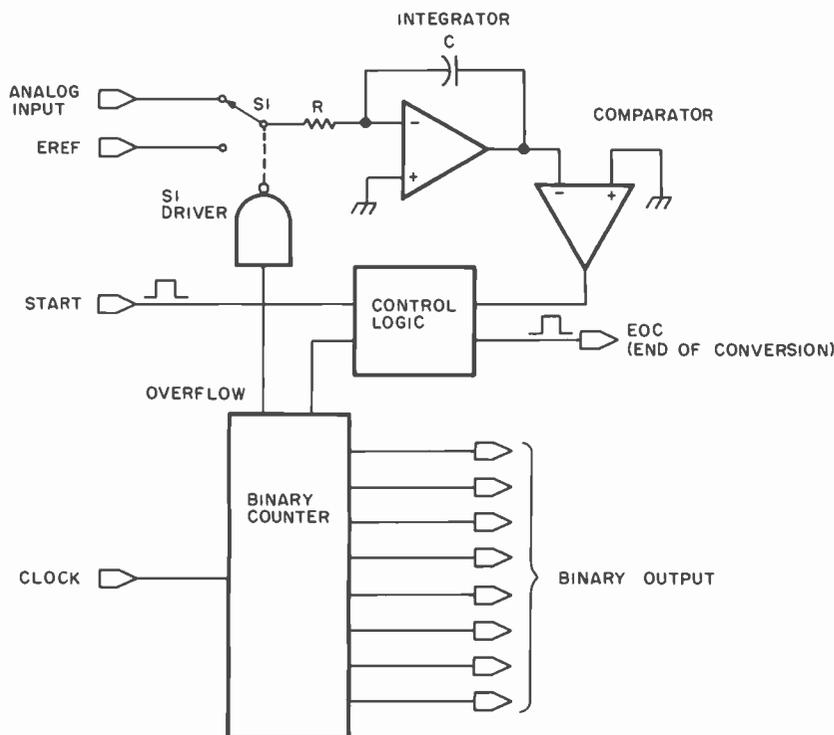
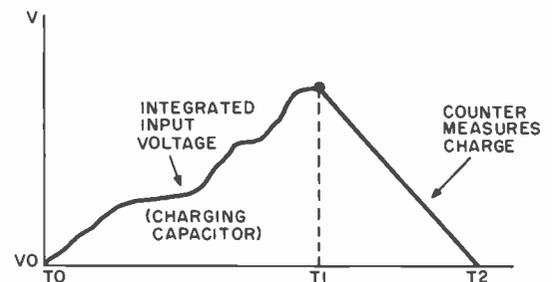


Figure 3: Analog to digital conversion by the dual slope integration method is often performed by slow devices such as digital panel meters. This method works through an analog integrator and a counter. The integrator has switchable inputs. It first integrates the incoming signal for a specified time interval. Then it counts the time necessary to linearly discharge the charge just accumulated with a known slope. The result is a count which is proportional to the voltage which drove the integrator during the charging time.



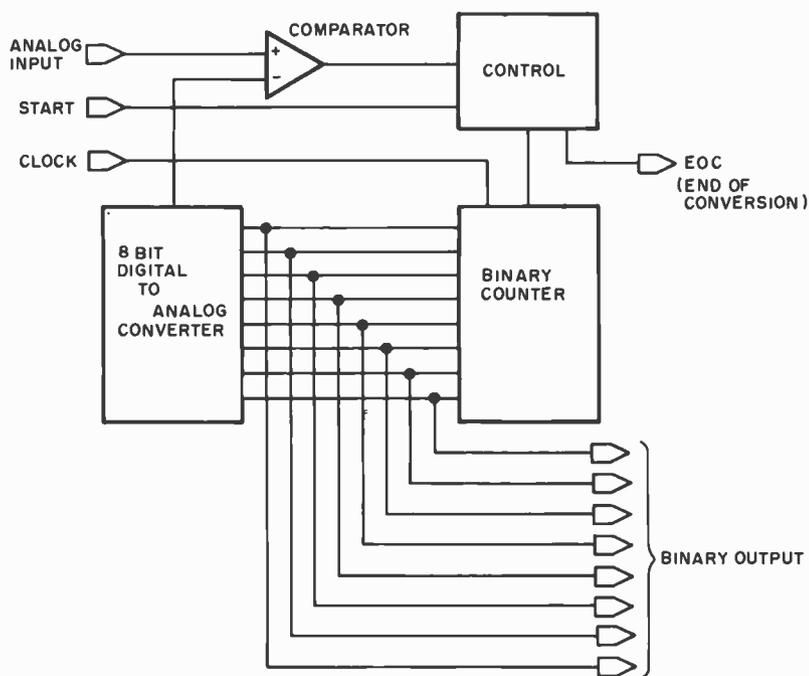


Figure 4: The ramp or counter method of analog to digital conversion is straightforward: A counter is initially zeroed and then allowed to count upwards until its binary code through a digital to analog conversion creates a voltage greater than or equal to the input voltage.

This makes its gain very high. If the two input voltages are not equal, then the operational amplifier output is high or saturated. In this case the comparator is ground referenced and uses just one active input.

When a START command is received the control circuit resets the counter to 00000000, resets the integrator to 0 V output (discharges C), and sets switch S1 to the analog input. The analog voltage creates an input current to the integrator which causes the integrator output to begin charging capacitor C. This means that the output voltage of the integrator begins to rise. As soon as this voltage rises a few millivolts above ground the comparator output snaps high. The high comparator output causes the control circuit to trigger the counter, which begins counting clock pulses. The counter is allowed to overflow and this outputs an overflow bit. This bit changes the state of switch S1. The graph in figure 3 shows the integrator charging during the interval between START and the overflow of the binary counter (t_0 to t_1). At time t_1 the switch changes the integrator input from the analog signal to a precision reference source. Also, at time t_1 , the counter has overflowed and again it has an output of 00000000 (maximum count + 1 is the same as the initial condition). It will, however, continue to increment so long as we have a high comparator output.

The charge accumulated on capacitor C during the first time interval is proportional to the average value of the analog input voltage between t_0 and t_1 .

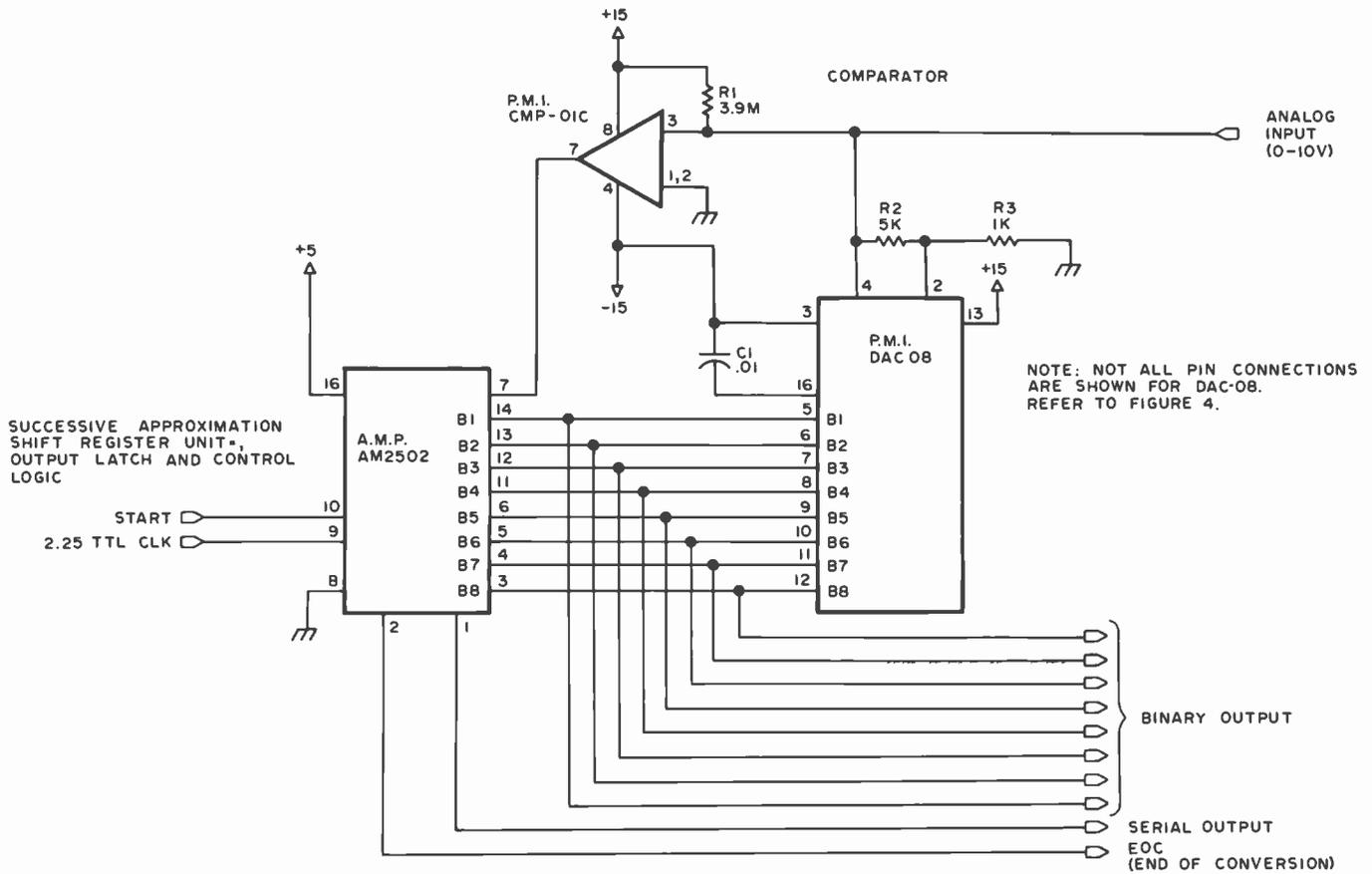
Capacitor C is discharged during the next time interval (t_1 to t_2). When C is fully discharged the comparator will see a ground condition on its input and again it will change state to make its output low. This causes the control circuit to stop the counter but *does not reset the counter*. The binary word at the counter output at the instant it is stopped is proportional to the average value of the analog waveform over the interval t_0 to t_1 . An end of conversion (EOC) signal is generated to let the microprocessor or other system know that the output data is stable, valid and ready for use. The speed limitations are based on the requirement for an accurate and stable analog integrator, and the need to average the input signal over a long cycle.

Counter (Ramp) Analog to Digital Conversion

A counter or ramp type analog to digital conversion circuit is shown in figure 4. Here we find a voltage comparator, a digital to analog converter with voltage output, a binary counter, and necessary digital control logic. Here is how the conversion works:

When the START command is issued by a control system (for example, a computer output port) the control logic resets the binary count to 00000000 and enables the clock input to the counter, which begins counting upwards at the clock rate. Since the counter outputs set the voltage level out of the digital to analog converter, the DAC generated voltage to the comparator will begin to rise. As long as the analog input voltage is greater than the reference voltage generated by the digital to analog conversion, the state of the comparator output will enable counting via the control logic. However, when the rising reference level finally equals or exceeds the input voltage for the first time, the comparator output state will switch and stop the counter. At this time, the output of the counter can be read by the computer or other system, and represents the value of the analog input voltage. If the counter and digital to analog converter are both eight bits, the number in the counter will represent from 0 (00000000) to 255/256 times the digital to analog converter's reference voltage level. The conversion time of this algorithm is proportional to the voltage being measured.

Both the dual slope integration technique and the counter technique discussed thus far take far too long for many applications. On the order of 2^n clock cycles are typically required where n is the number of bits involved. Conversion time becomes critical in an application when the frequency



response must be high and a faithful reproduction is required. (For reference, consult textbooks in electrical engineering concerning the "sampling theorem" and Nyquist's criterion that says we must have a sampling rate of at least twice the highest frequency that is to be recognized, if a faithful reconstruction of the signal is to be achieved.)

Successive Approximation

In programming and software design, we often find use of a "binary search" stratagem instead of a flat out sequential search when trying to speed up the process of finding an item in a table. This same approach is just as applicable in hardware, where the "successive approximation" technique of analog to digital conversion provides a much higher speed of conversion than the sequential counting methods discussed so far. The successive approximation technique typically requires only $n+1$ clock cycles to make an n bit conversion, and its hardware is no more complex than that of the dual slope or counter methods.

The successive approximation converter, shown in figure 5, consists of a comparator, control logic, a shift register with output latches for this form of conversion, and a voltage output digital to analog converter.

When a START command is issued to the converter circuit, it loads a binary 1 value into the most significant bit of the shift register, which in turn sets the most significant bit of the output latch. This sets the output of the digital to analog converter to half scale. In true binary search strategy, if the input voltage is less than the reference output provided by this half scale setting, the most significant bit is cleared from the latch on the next clock pulse; otherwise, the most significant bit is left unchanged at the next clock pulse. Then the internal shift register of the successive approximation register unit is shifted so that its single high level bit is opposite the next most significant bit. Again, the output register is modified, this time so that bit 2 is set to 1 for the trial measurement. This bit has a value of one fourth of the total voltage range, which is added to the half range or zero value still latched from the first measurement. At the next clock pulse, if this new trial value to the comparator is greater than the input value, the 1 bit is latched in the successive approximation register; otherwise a 0 bit is inserted at the current position. This process continues with successively less significant bits until the shift register overflows indicating that the last bit has been tested. Some forms of this equality have control logic to detect an equality condition and

Figure 5: The successive approximation method, here illustrated with a practical circuit, uses a binary search strategy. The most significant bit is tested first, then the next most significant bit, and so on down the line until all n bits have been tested. If at each stage the contribution of the selected bit causes the trial approximation output from the converter to exceed the input value, the bit will be stored as a zero. After all n bits have been tested, the result is an n bit binary representation of the voltage of the input.

Continued on page 62

Ask BYTE

BRITISH COMMENTARY

As one of the relatively small number of personal computing addicts in England, I'm very impressed with the rate at which the field is growing in popularity on the other side of the Atlantic. I've been subscribing to BYTE for the past six months and I promise it's always made fascinating reading. I wish I knew where in this country one could lay hands on any issues before last August's. (Hint, anybody?)

The advertisements are fascinating, too. Anybody considering purchasing any sort of system is almost forced to import it from the USA. My query is about TV displays that most of these systems use. Naturally, they're designed for American TV standards, which differ from British ones. We have 625 interlaced lines per frame, repeating at 25 Hz. Is it, in general, possible to modify an American TV display device to work with a British TV? Or is all the timing generation usually performed inside a special chip? Might I have more success if I tried to modify a TV instead? I know rather more about TTL logic than about televisions, but I'm game to learn if necessary. Otherwise, if the answer to all these questions is no, then the temptation to come and live in the States is going to be almost unbearable

I look forward to every issue of your magazine; I'm only sorry that I discovered it so late. Best wishes for the successful future that you deserve.

Guy A Burkill
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Cambridge CB2 1RH ENGLAND

Are there any other British readers who have experienced the problems of interfacing American video generation logic to European television designs? Some firsthand knowledge of the subtleties to be found would be the best way to answer Guy's query on that subject.

With regard to back issues of BYTE, there are none left. However, much of the editorial content is now available in a book titled Best of BYTE edited by David Ahl, publisher of Creative Computing. This book contains reprints of numerous articles from the first 16 issues of BYTE, September 1975 thru December 1976. ■

SACRED BUSES

I am very curious about something. What is sacred about the Altair bus as opposed to others? Would it not be possible to install a peripheral designed for the Altair scheme to, for example, a

Digital Group or Southwest Technical Products bus? I am confused on the issue since it seems to me that, functionally, lines must correspond between the systems pretty well. Perhaps the answer lies in the fact that the Altair bus was based on the 8080 processor, which has some unique IO methods. Is that the problem? Is it really a question of 8080 versus other processor compatibility? I understand why a manufacturer would want to make plug compatible cards to the Altair bus, but why couldn't a simple conversion be available for any product?

J C Chirigos
1601 Kentucky NE
Albuquerque NM 87110

The main issue is one of plug-in compatibility. Whether you call it the Altair bus, or, as used by non-MITS suppliers, the S-100 bus, the key to the wide availability of the peripherals is nominal compatibility at the hardware level. Even here, there are occasional clashes between various manufacturers about the definitions of pins not originally assigned meanings in the Altair definition of the bus.

At the detail level, a 100 pin bus surely works, and in principal one could talk to any other similar computer at the memory and bus interface level simply by simulating all the signals which would have been present on the bus in its Altair implementation. This is a quite workable procedure, as demonstrated by many products.

True, the IO structure of the 8080 is unique, and there are some 8080 specific features of the Altair bus as a result of this and other characteristics of the 8080. But for the most part, the particular selection of the lines present on the 100 pin interface of the Altair bus is just a reflection of discretionary choices on the part of the designer(s) of the first Altair within the framework of the general design of a microprocessor with 16 address bits and eight data bits. ■

SOME TERMINOLOGY

I have a problem.

First, let me give you some idea of my background. I have spent 15 years in the computer industry as an applications programmer. I am fluent in BASIC, FORTRAN, PL/I and APL. Now that I have left the industry, I am getting interested in recreational computing.

I have never been involved in logic design or in the details of hardware, nor do I feel that I want to get involved with it now. I would not mind building a kit provided that the instructions were of the "put tab A into slot B" type. However, when I read all the ads and literature furnished by the various manufacturers, I have the following problem: What are you all talking about?

I thought I knew what "read only memory" was. Now I come across "pro-

Continued on page 67

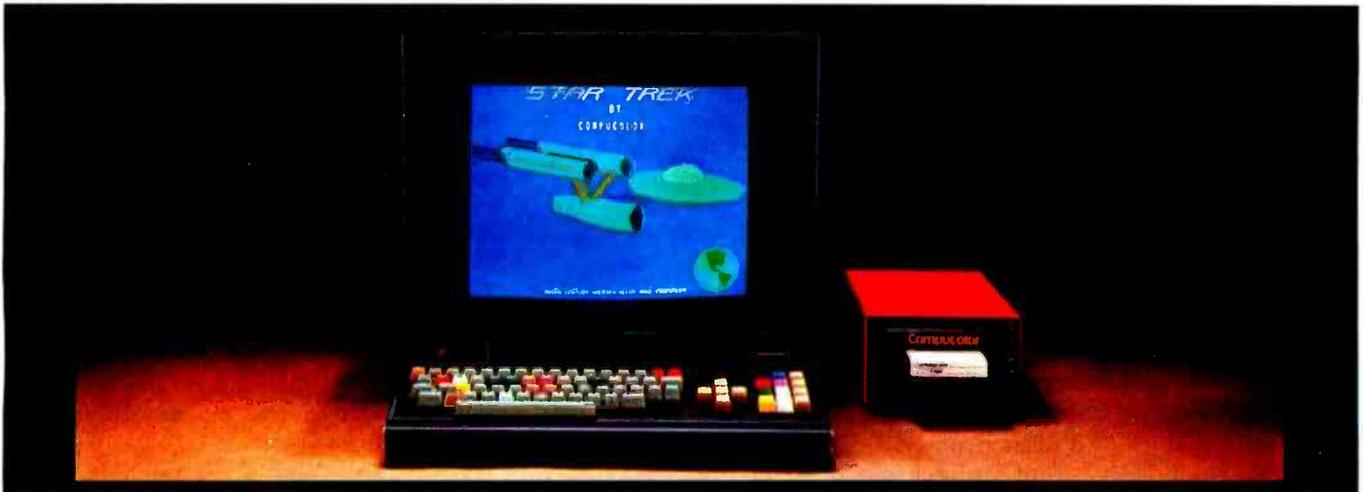
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stop the conversion ahead of overflow, but the worst case time is the n+1 clock pulses for n bits mentioned earlier. Figure 6 shows the sequence of voltages presented to the comparator by a 6 bit successive approximation algorithm, as compared to a 6 bit counter conversion algorithm to show the time savings of this method.

In the circuit of figure 5, the comparator is shown as the Precision Monolithics CMP-01C device, although faster conversions may be possible if a higher speed Advanced Micro Devices AM686 comparator is used instead. The Advanced Micro Devices AM2502 integrated circuit, which was designed for this successive approximations conversion application, contains everything needed for the logic described verbally above, except the digital to analog converter and the comparator. Other companies go even further with integration of the input conversion. Precision Monolithics, for example, makes an AD-02 circuit which contains all the complete 8 bit analog to digital conversions. It is relatively expensive, but its cost can often be justified by its utility and ease of use. It has a respectable conversion speed and has several input options that can accommodate analog voltage ranges of 0 to 5 V, 0 to 10 V, -2.5 to +2.5 V, -5 to +5 V, and -10 to +10 V.

Software Approaches . . .

As noted earlier, the software of a microprocessor can often implement the algorithms of digital to analog conversion. This is especially so with the successive approximation algorithm, since its inherent speed makes up for some of the slow facts of life concerning programmed execution. To rig a software approach to the problem, we need a digital to analog converter attached to an output port, an input comparator which drives one input line, and the software of successive approximation (or other methods for that matter). In this case, the successive approximation shift register is variable in a program, the output latch is an output latch connected to the DAC device, and decisions are made based on the single bit input. A previous article in *BYTE* [see "Microprocessor Based Analog/Digital Conversion," by Roger Frank, page 70 of May 1976 *BYTE*] discussed both the ramp (counter) and successive approximation methods described here, but showed how to implement them in software.

Whether the approach taken is that of pure hardware or software aided designs, adding analog input conversions to a personal system can expand its capabilities to cover many interesting real world control and measurement problems. ■

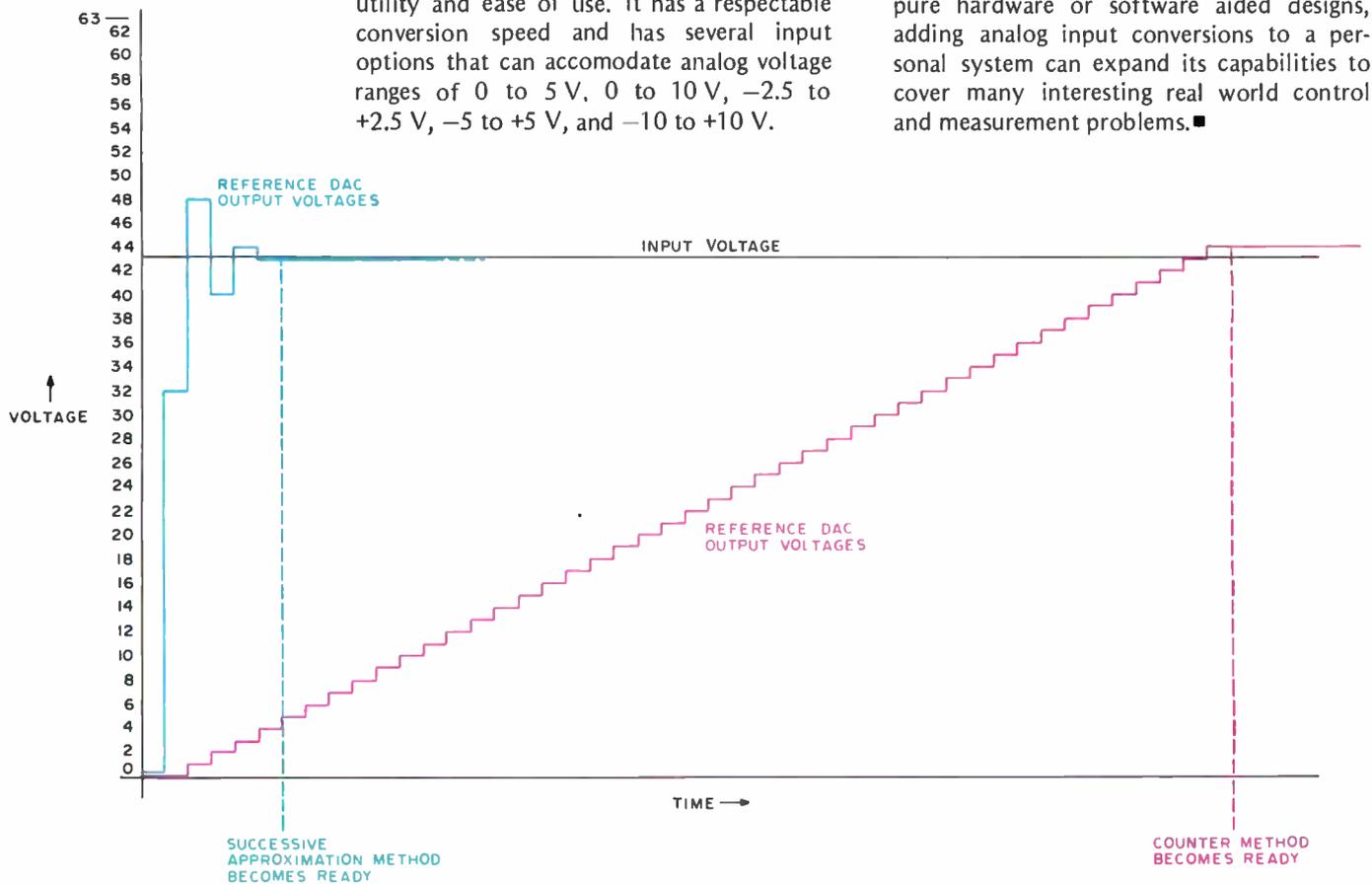


Figure 6: Comparison of the successive approximation method with the ramp method of conversion for a 6 bit value. (Six is chosen for purposes of this illustration.) The successive approximation method takes a mere six cycles of testing to arrive at the best value, where the ramp method has to count up to the number which matches the input and causes the conversion to terminate. The ramp takes 44 cycles here, versus six for the successive approximation method.

*Personal Computing
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Personal Computing Program

Two full days of panel sessions on June 15 and 16 will provide an in-depth look at Personal Computing: Past, Present and Future; The Future of Retail Computer Stores; Hardware of the Computer "Hobby" Market; and Personal Computing Software. Leaders in the Personal Computing movement will appear on each of the panels to let you know the latest developments in the field, point out trends you'll need to be aware of—and answer your questions.

Special Interest Sessions

In addition to the panel sessions, special interest groups will be able to gather informally for "how-to" programs on building a kit, debugging software, using assembly language, I/O interfaces, cassettes and disks, software standards and so on, into the night. If the special interest group you want is not organized when you get there, we'll do our best to help you get one started!

National Club Congress

Is a national personal computing association needed? If it is, what does it do, how does it do it, and who does it? To find out what's happening—pro and con—club reps from across the nation will gather to exchange ideas and discuss issues related to club activities and programs. Make certain your club sends an official delegate who can speak for you and vote vis-a-vis a national organization, establishment of national hardware/software standards, a national program library and interchange, educational seminars, meetings, *ad infinitum*.



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Special plenary sessions will feature a keynote address June 13 by Mark Shepherd, Jr., chairman and chief executive of Texas Instruments Inc.; the AFIPS Presidential Address June 14 by Dr. Theodore J. Williams; and a special address June 15 by A. Douglas Murch, senior vice president, Prudential Insurance Company of America.



Other highlights will include a Pioneer Day Program honoring members of the computing group at Los Alamos Scientific Laboratory; the first NCC National Programming Contest; the annual NCC Computer Science Film Theatre; special tours and an all-conference reception.



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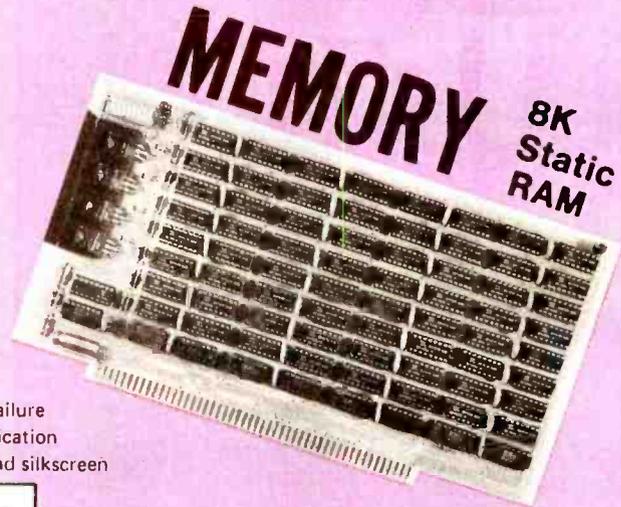
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Continued from page 60

grammable read only memory" (or worse yet, erasable programmable . . .). If it's read only, how can it be programmable? What is static memory? Is there any other kind of memory other than "random access memory?"

Is it possible for you to help me out of my dilemma? Something between "binary numbers are made up of zeros and ones" and "when the static EPROM is connected to the DMA using a 3P+S IO module."

Al Weiss
POB 942
Alleghany CA 95910

Short of the tutorial article which may be inspired by your questions, here are a few notes on terminology and concepts in computer design and implementation.

Random Access Memory: For the purposes of discussing the present state of technology, a random access memory means some sort of reference to semiconductor memory parts (excluding serial access devices such as shift registers). A magnetic core memory is a form of random access memory, but is not used in modern small computer technology due to various manufacturing and

economic considerations. A read only memory (ROM) is a random access memory part, as is an ordinary programmable memory. The read only memory is distinguished from the fully programmable memory by the fact that it is non-volatile (the information is retained when power is removed) and moderately difficult or impossible to alter once it has been set up with a program.

A commonly used acronym found in technical jargon and advertising is RAM, referring to random volatile programmable semiconductor memories. Since both read only memories and volatile programmable memories are random access devices, this term is misleading and ambiguous. In BYTE's editorial content, we do not use the term RAM intentionally, because of this ambiguity, and refer instead to "programmable" and "read only" memory. Programmable emphasized the volatile, user program oriented nature of parts which are often called RAMs in conventional engineering journals, and read only characterizes the nonvolatile, permanent nature of the other type of random access memory part. (Of course, "programmable" is still not an optimal choice, since even read only memories are always programmable in the sense of "program it only once.")

Static Versus Dynamic: In brief, there are two types of volatile programmable memory parts, characterized by the in-

ternal design of the basic memory cell of the circuit. A static memory typically has a cell with sufficient active transistor elements to create a true flip flop memory register. A dynamic memory typically has a smaller memory element size which is achieved by replacing the memory flip flop with a capacitor which stores an electric charge. As a result of this smaller unit bit cell size, at the limits of technology the largest dynamic parts have historically had about four times as many bits as the largest static memory parts. At the present time, the largest static memories readily available are 4 K bits per chip, whereas the largest dynamic memories are 16 K bits per chip.

There is a subtlety of design with dynamic memories, however. This is the fact that since the storage elements are capacitive in nature, sensing and support electronics on the chip tend to drain the charge with time, losing any information stored in the cell. The dynamic memory chips must thus be refreshed periodically, an operation that is commonly performed by cycling through a reference to the low order bits of the address inputs to the chip. Static memory chips have no such refresh requirement, and are often easier to use in prototype, small or homebrew circuitry; the manufacturing economies tend to make dynamic memories the most attractive in larger systems products. ■

Think about it!

If you could design your own computer system from scratch, you'd do it right.

You'd want...



Continued from page 9

number of existing computer science textbooks. But I'd really like to somehow buy the design in a completely documented form so that like the plan of the hypothetical rolltop desk, I could implement it literally with custom modifications. Tutorial and "how to" plans books for specialized fields such as those mentioned above are widely available already, and at prices well within the range of an individual's budget. They are marketed in large quantities because large numbers of individuals use the information; outlets range from mail order book services to retail stores. Drawing out of this parallel between individualized computing and individualized "anything else," it should be obvious what the solution of the software dilemma is: Publish detailed plans and tutorial information for software, on a scale commensurate with the size of the market. Publishing ideas is an activity which has a long and distinguished history, and yields both personal and financial rewards to those who engage in the practice, as well as real benefits to those who purchase the products. Let's turn now to the application of this concept to the software designs of the computing world.

The Ideal Model of a Published Software Product

When we talk about publishing software at the present state of technology, we are talking about a product which is akin to the detailed design of the rolltop desk mentioned earlier. It is a product which serves as the starting point for the home software craftsman, not a recipe which will fit without thought into every conceivable system. This will change a bit as the systems in the marketplace become more refined, but the nature of the computer as an intellectual amplifier tends to require a certain level of technical familiarity on the part of its user. (This is the element which distinguishes the general purpose computer from the applications oriented dedicated computer such as a 4 function calculator or oven controller.)

In order to make a software package which is optimally configured for the customer's standard or customized use, there is a certain minimal level of documentation which is required. This level of documentation is not necessarily needed by all users all of the time, but is in many respects akin to the reference books for integrated circuits: When a question needs to be answered, it is good to have the information needed to zero

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in on the answer. Here is what I consider to be adequate documentation:

- Users' manual textual materials concerning the "standard" uses and limitations of the software. Here is where we find such information as standard IO patch points, relocation tables, etc.
- Complete object code, preferably machine readable along with machine readable relocation information.
- Complete source listing of the package including source language and generated object code for each statement.
- Program logic manuals and tutorials on the design of the product are an excellent option.

The idea is to include enough information to allow the user to do routine field alterations, including relocation. The idea of a published software product is to compile all this information together in a comprehensive book form, to be sold at prices characteristic of books, as opposed to the past history of software prices for applications and system software packages. The technology of printing covers all the portions of the "complete" package except machine readable code, at least in the minds of most people. However, as we have demonstrated with experiments

published in BYTE, printing technology also covers machine readable representations as well.

Varieties of Machine Readable Representations

User convenience demands that a software product be made available in some form of machine readable representation. While it is certainly possible to take an object listing in printed form and type it into a processor by hand, this is a long, tedious and error prone process. To complete the functional definition of "adequate documentation" given above, we need a form of machine readable object code at minimum, along with machine readable relocation information. Fortunately there exist several technologies which can be employed for this purpose, which I'll review here.

ROM Releases

This is the most expensive medium presently available for reproducing software; however, it has utility in the convenience of use provided by built-in software. In terms of practical products, however, this form of software will most frequently show up in manufactured products preloaded at the

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factory, rather than user integrated software. A ROM program is difficult to achieve in a relocatable form, and has a certain permanence which is both its advantage (convenience) and its disadvantage (difficulty or impossibility of patching). This method has been successfully employed in several desk top calculator packages in the form of ROM software options, and in such personal computing products as built-in BASIC interpreters and monitors. For end user markets, this form of software can dispense with all but the user oriented manuals in most purchases, since modification is impossible.

Magnetic Digital Media Releases

As more and more floppy disk products, large and small, come to market, the use of the magnetic diskettes for software releases is becoming common. Similarly, the Philips and 3M digital tape media with standard digital recording techniques can be considered as vehicles for release of machine readable data. However, the price of the media and the costs of digitally recording and verifying each copy tend to make this method of delivery limited. It is used, quite naturally, as the vehicle for delivery of floppy disk operating systems from the

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manufacturers of drive interfaces, but there the writing and testing of a diskette full of data falls out of the expected quality assurance tests prior to packing and delivery. The same is true for the other forms of hardware which have related operating system software products that can be recorded.

Paper Tape

This venerable medium has been in existence longer than the modern electronic computer. No survey of distribution media would be complete without mention of it. Very reliable means exist for reading paper tapes into computers at quite economical prices, well under \$100 for the peripheral. As a distribution medium, however, paper tape in my opinion suffers from several disadvantages: It is inconvenient to store, bulky, prone to create a messy tangle due to manual handling with inexpensive peripherals. Whether my opinion is supported in the marketplace is another question altogether.

Audio Media Releases

One of the most useful and practical vehicles for the distribution of software is likely to be the use of audio recording media. Here we can identify two principal

methods of distribution; recording on tape cassettes or other magnetic tape audio media, and recording on the audio equivalent of a read only memory, the phonograph record.

The technology of recording on tape results in a product with a fairly high unit cost for each copy of the information. To this must be added the cost factors associated with normal printing of the rest of the documentation. Cassettes, as a recording method, are a logical choice for custom software, or small volume situations, but the high degree of manual labor associated with each copy argues against the practicality of large production runs in this form.

The technology of making phonograph records is, on the other hand, a well established mass production technique which can be adapted to the software distribution without much variation from standard methods. To illustrate the point and to test the concept, I made a test in the spring of 1976 at the suggestion of David Fylstra, an associate of mine who is also a homebrew record maker. He arranged for the cutting of a test record with the audio format of my personal monitor program, circa March 1976, to test out this method using a master record cut on standard recording industry equipment. Depending upon size and quan-

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tity of pressing, the costs per record run well under \$1, which is hard to meet with the cassette duplication method.

Machine Readable Printed Media

As a final option, there is the use of machine readable printed formats for object code and relocation information in the release of software products. This is a form which was suggested to us at BYTE by Walter Banks of the University of Waterloo, and with which we have been experimenting in the pages of BYTE. In this method, an optical reader is used to scan printed materials which have been formatted into a series of bars corresponding to the digital information. Because of constraints in the design of the layout and the method of scanning, it is possible to simplify the scanner designs to the point where a very inexpensive peripheral is used together with some adaptive software which takes care of the speed tolerant input scan. The beauty of this method is that it "comes for free" in so far as actual production costs are concerned. Why is this true? The reason is that the 200 to 300 pages of documentation needed to support a systems software product with perhaps 12 K bytes of object code require only an additional five to seven pages of

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machine readable bar code copy, hardly affecting the economics of the book at all. These 200 to 300 pages of documentation are required by the product concept, whether or not there is any other form of machine readable code made available.

Economics of Publishing

With media established, and a product concept outlined, what about addressing the problem of rewarding the producers of software products? Here, as in any area of publishing, the answer is quite simple. The publishing house judges whether the particular software package is in its view a readily marketable product with a certain minimum press run potential. If so, the publisher puts up production capital, where the author puts up intellectual capital in the form of his or her work. It is a risk situation in which both parties are making a speculation that readers will purchase the product; as in numerous parallel situations throughout industry, authors and publishers work on an agreed upon split of any rewards from success in the marketplace.

Applied to the software publishing variant of this business, the author's intellectual capital is in the form of the program, its source code, its object code, and its documentation; the publisher's contribution is the marketing organization, the technical editing of the manuscripts, and the technical details of book preparation. Other than the specialized content, the method of operation and the details of the arrangement are not much different from publishing any item. Rewards to authors now become a small

amount (in absolute terms) of royalty recovered from orders of magnitude in sales for successful software book products.

Proprietary Products

The problem of protecting and keeping software proprietary is no longer a major "new" issue when publishing of software is contemplated. How many people extensively copy from books? Very few, and if they attempt to make a regular practice of it they would tend to be prosecuted by publishers under copyright law. In publishing software, an implicit or explicit license to copy the copyrighted materials for personal use and modification is part of the bargain; the price is low enough so that if you want your own user documentation, you buy your own copy of the book (even if you may have been using object code derived from your neighbor's computer). Since the documentation is a necessary component of use, no sales tend to be lost in the long run due to the fact that object code can be swapped around.

Conclusions

What I have endeavored to show is that there is quite some potential for the sale and distribution of software using conventional publishing techniques with modifications to suit this type of product. By publishing software along with machine readable code, we end up with a way to make the products widely available, yet retain the desirability of compensating authors for their efforts in proportion to the success of the product. ■

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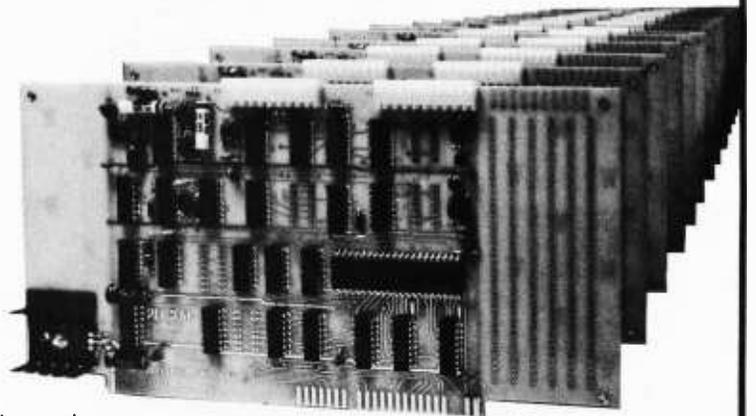
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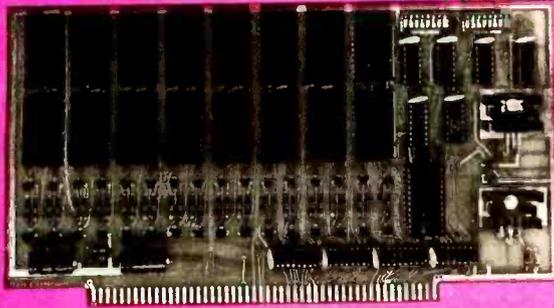
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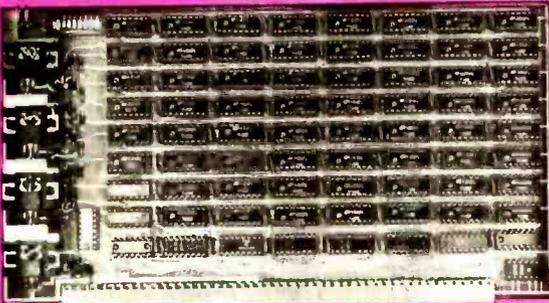
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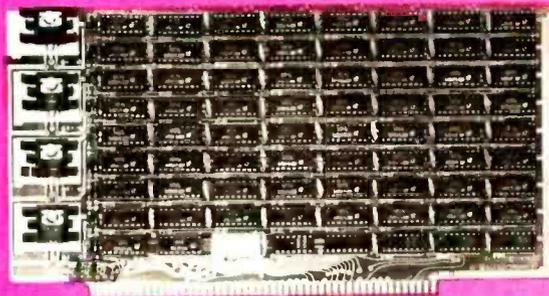
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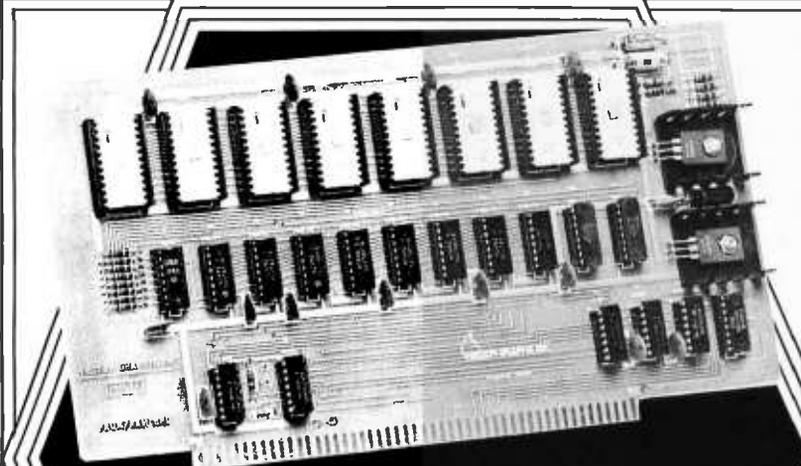
A short note concerning your note in the March 1977 issue of BYTE (page 100) on the Signetics 8X300. This processor was designed and sold as the SMS 300 (Scientific Micro Systems, 520 Clyde Av, Mountain View CA 94043). I used SMS's development system in a scientific data acquisition system. Two years ago it was the fastest thing on the market. I was building a two dimensional multichannel analyzer out of a Modcomp II, and used the microcontroller for the address mapping and handshaking. Pure blinding speed was our object, subject to the constraint of programmability. The observation is that this is the same criterion for a microprocessor used for emulation.

True, the 8X300 is fast, but by today's standards, not that fast. It is rumored that

SMS is working on an ECL version Anyway, if one examines the architecture of the 8X300, one finds two chokes, both fixable. The first is that the microcode is quite vertical, as one would expect with a 16 bit instruction. One can effectively double the speed of the system by simply extending the microcode width by eight bits. Make your program memory 24 bits wide; use 16 in the normal fashion. The extra eight are used as the address in working storage or IO space. The 8X300 must use an instruction to load the "memory address register" (IVR REG) and then another instruction to fetch the word at that address; by selecting an address for each instruction (the extra eight bits), most programs halve in size, and double in speed.

The second choke is the time multiplexed IO bus. I seriously suggest the user extend the microcode and stay away from those

Continued on page 110



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Photo 1: The Teleterminal Corporation Fly Reader for use with the KIM-1 microprocessor.

Come Fly With KIM

Rick Simpson
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Many computer hobbyists start with nothing more than a processor, a small amount of programmable memory, a small onboard monitor such as MIKBUG or KIM and some front panel switches. Those with more foresight, or cash, will have a keypad or even a full keyboard for data entry and processor control. But even with a good monitor and a full keyboard and display, loading programs is a tedious chore at best, and there is an awful feeling when you turn off power, knowing that twenty minutes of typing just evaporated.

The next step in expanding the system is usually an audio cassette interface or a Teletype with paper tape reader and punch for the wealthy or fortunate. Now the tedious retyping is eliminated and a program, once written and recorded or punched, can be reloaded in a matter of minutes.

Many people stop at this point. When hand assembly of programs is required, a program of more than a few hundred bytes is rarely attempted. But as the software gap is slowly filled, more and more systems are being implemented with assemblers or

BASIC interpreters. More memory is purchased to expand programmable memory from a few hundred bytes to 4 K, 8 K, or more. *[One firm now even markets a 64 K board! . . . CH]* Once again your memory has outrun your ability to fill it in a reasonable time.

For instance, using the Teletype paper tape reader or audio cassette interface on the KIM system, a 2 K Tiny BASIC interpreter takes almost ten minutes to load. A 12 K BASIC source program would take an hour. Even a 30 character per second interface only cuts this to twenty minutes. The alternatives seem to be a Tarbell or Suding type high speed cassette system, a 3M drive at 9600 bps or a floppy disk.

The floppy disk certainly solves the speed problem. We can now load 12 K in a few seconds, but at a cost of \$1,000 to \$2,000. The high speed cassette is reasonable in cost, about \$200 including the high quality cassette unit required, but tricky to interface unless a manufacturer-supplied board or kit is available.

Although some magnetic tape units have start, stop, and search functions under pro-

gram control, most users end up pushing the buttons. No hobbyist magnetic tape cassette unit can read a few (ie: one line or so) characters, stop and process the data, and then start and read some more, a real need when running an assembler with the source stored on the tape in a limited resource system.

After this lengthy preamble, you may have suspected that I have an alternative solution in mind, and I do: a high speed paper tape reader, manufactured by the Teleterminal Corporation, called the Fly Reader, shown in photo 1. Although a bit more expensive than the high speed cassette system (about \$350), it is far faster; reading at 300 characters per second, it can load my Tiny BASIC in twenty seconds, or fill that 12 K of memory in two minutes. It is easy to interface, requires little software, and is extremely reliable. It needs only a single +5 V, 2 A power supply and is operated completely under program control. You can read as little as a single character at a time and can read in either direction; try that on your cassette!

Paper tape has always been the standard mass storage device for minicomputers, until floppy disks came along, and paper tape has been the most universal and inexpensive method of software distribution and interchange in the minicomputer field.

The basic problem is that it is only a reader; how do you punch the tape? There are several answers: Flexowriters and other similar low speed punches are becoming available, as are gobs of older 7 level machines. I've also seen higher speed punches, typically 60 characters per second, advertised for under \$100. The fact that the punch is slow is not so important; typically you punch a tape once and read it many

times. Even if you have no punch, the reader is a useful peripheral because much software is available already punched.

How it Works

The Fly Reader can read at such a high speed because it transfers 8 bits in parallel and contains only a single moving part: a stepping motor connected to a toothed wheel which engages the sprocket holes in the paper tape. Sensing of the holes in the tape is done by photodetectors rather than the mechanical fingers used in a low speed reader. This is a method similar to that used in the manual reader sold by Oliver Audio. Figure 1 shows a block diagram of the unit.

There are five control lines for the unit. All are compatible with standard TTL circuitry. The "load status" line is a logic 1, +5 V, if the reader is not ready because the feed gate is not closed. When tape is inserted and the gate is closed, this signal goes to logic zero.

In operation, the reader must be checked by software to see if the "reader ready" signal is at logic 1 to indicate that the reader is ready to read another tape character. The software must then issue a pulse from logic 1 to logic 0 whose width is between 500 ns and 500 μ s. This READ pulse will start the reader and drive the reader ready line to logic 0. The software then watches the "data strobe" line. When data strobe goes to logic 1, the data can be read from the eight parallel output lines. If the program needs to read another character from tape, it must wait until reader ready goes back to logic 1, issue another read pulse, and wait for another data strobe. Figure 2 shows the flowchart for such software and figure 3 shows the interface timing diagrams.

Figure 1: Block diagram of the Fly Reader. The input is achieved through an incandescent lamp and a photodetector array. The tape is advanced by a stepping motor allowing input of data either forwards or backwards. The reader open sense switch is closed when the paper tape is in the reader.

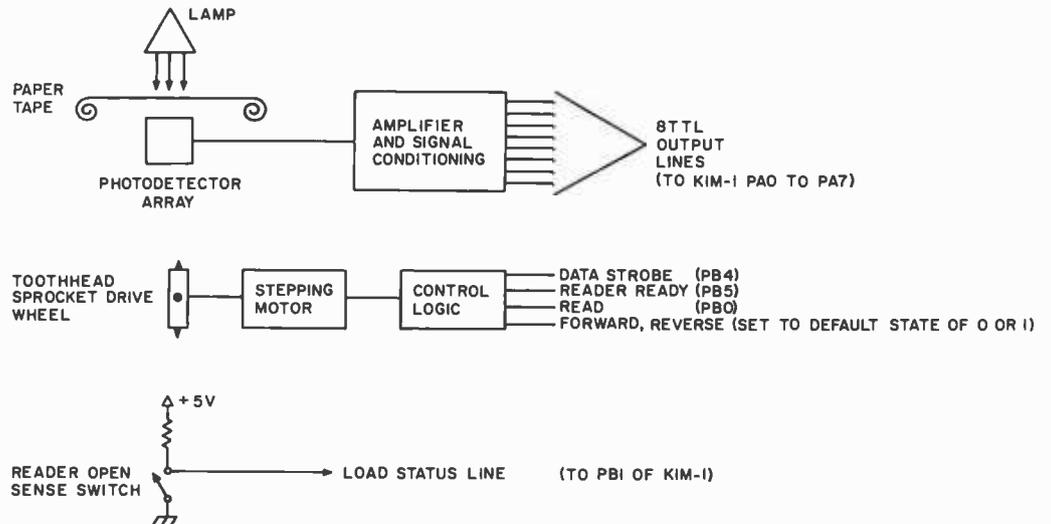
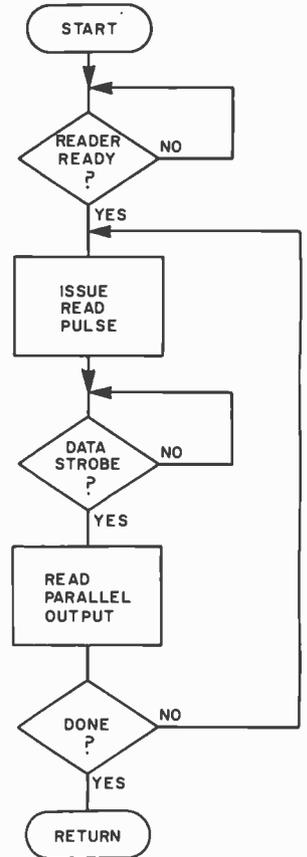


Figure 2: Flowchart of the software for reading the paper tape with the Fly Reader.



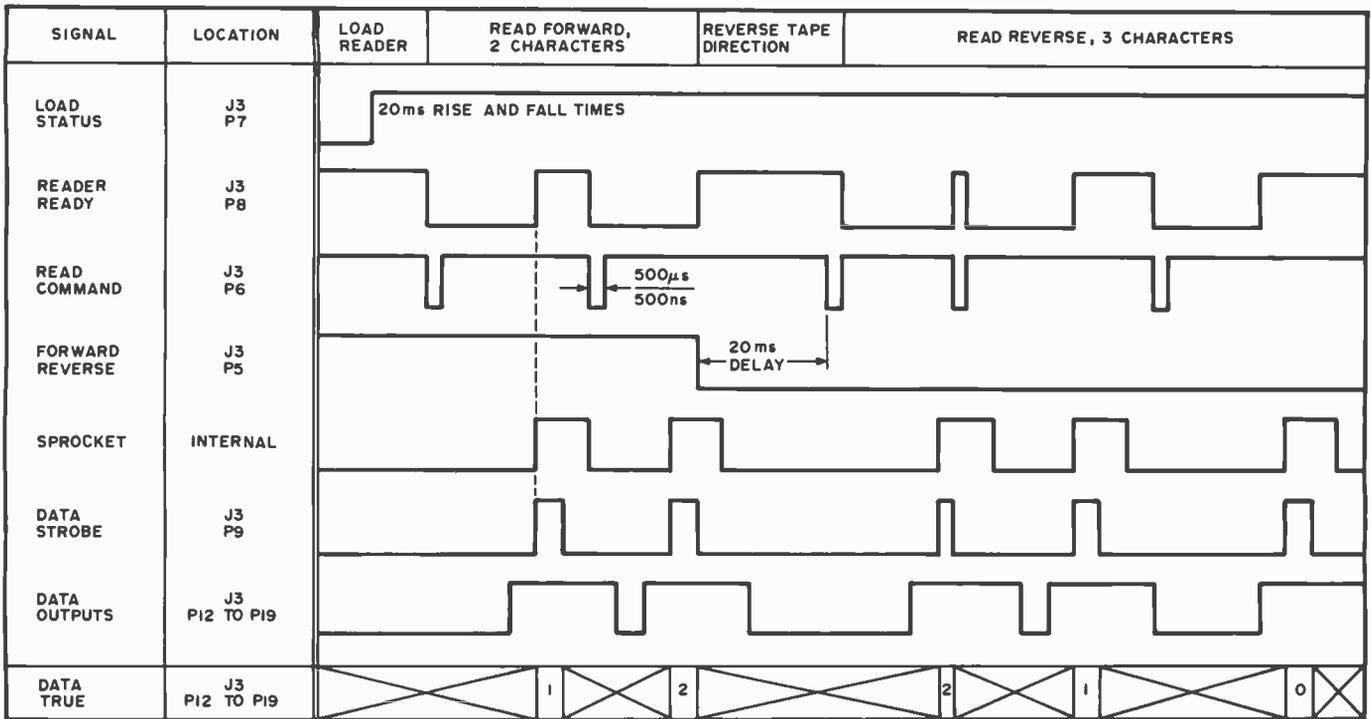


Figure 3: Timing diagram generated by the software of listing 1. The minimum width of the data strobe is 50 µs except when forced low by a new read command. The crossed out sections in the data true section indicate that the state of the output is unknown. A read command is issued only when the reader ready line is high.

Interfacing to Kim

As described above, the Fly Reader interface requires eight parallel input lines, three input control lines and one or two output control lines. Two output control lines are needed if the forward, reverse function is used; otherwise only one control line is needed. Since KIM-1 has 15 bidirectional IO lines the interface is very simple. The A data port lines PA0 to PA7 are programmed as input lines and connected to the parallel output lines from the Fly Reader. PB5 is connected to the reader ready line, PB1 is connected to the load status line, PB0 is connected to the read command line and programmed as an output line, and PB4 is connected to the data strobe line. A 5 V, 2 A power supply is connected to the reader and the interface is complete. When wiring the power connectors, you should make sure that separate power and ground wires are run back to the power supply for both the motor and logic connections. This will insure that current surges to the motor during stepping operations do not feed noise pulses into the control logic.

Interface Details

The fifteen KIM-1 IO lines are divided

into two ports. Each port has a data direction register and a data register. Writing a 1 to a bit or bits in the data direction register configures the corresponding IO lines for output, writing a 0 sets them for input. For instance, writing a hexadecimal 02 to the A data direction register configures the PA0 line for input, PA1 for output, and PA2 through PA7 as input lines. Similarly, writing hexadecimal F0 to the B data direction register configures PB0 through PB3 as input lines and PB4, PB5 and PB7 as output lines. Note that there is no PB6, and PB7 has no output pullup; it is essentially an open collector output. Reading the A or B data register will show whether the signal at each input line is 1 or 0 and will show whether a 0 or 1 was previously written to any lines configured as outputs. Writing to a data register will set the appropriate output lines to 1 or 0 and does not affect lines programmed as inputs. Hexadecimal address location 1700 is the A port data direction register, hexadecimal 1702 is the B data direction register, hexadecimal 1701 is the A data register, and hexadecimal 1703 is the B data register.

The KIM Paper Tape Format

The software to drive the reader uses the same paper tape format as that used in the KIM-1 Q (paper tape dump) and L (paper tape load) commands. Thus any paper tape punched on a low speed punch by KIM-1 can be read by the Fly Reader. The KIM-1 paper tape format ignores any characters

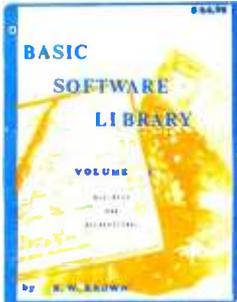
Other Interface Possibilities (An Advertisement . . .)

The interface described here uses most of the available IO lines on the KIM-1. Systems supporting several IO devices may wish to interface the Fly Reader through a separate interface chip. The MCS6532 is one such interface chip for adding more IO to KIM-1 as well as an additional 128 bytes of programmable memory and another interval timer. Two of these chips (with ROMs) are already built into each KIM. Since the 6532 is a MOS rather than TTL device it does not load the KIM-1 address or data buses significantly. The MCS6532 is available for \$16 postpaid from National Electro-Sales, 12063 W Jefferson, Culver City CA 90230.

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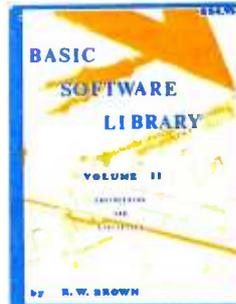
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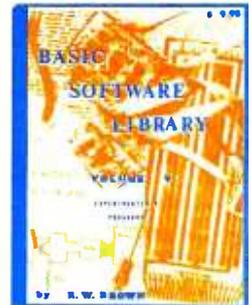
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Each program's source code is listed in full detail. These source code listings are not reduced in size but are shown full size for increased readability. Almost every program is self instructing and prompts the user with all required running data. Immediately following the source code listing for most of the programs is a sample executed run of the program.



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4007	C9 38		CMP	\$3B	} if A not equal to 3B go to LOAD;
4009	D0 F9		BNE	\$LOAD	
400B	A9 00	LOADS	LDA	00	else A:=00;
400D	85 F7		STA	\$F7	} store checksum;
400F	85 F6		STA	\$F6	
4011	20 8B 40		JSR	PTRBYT	go to PTRBYT; [get byte count]
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4020	20 8B 40		JSR	PTRBYT	get low address pointer;
4023	85 FA		STA	FA	store low address pointer;
4025	20 91 1F		JSR	\$1F91	compute checksum;
4028	8A		TXA		A:=X;
4029	F0 0F		BEQ	LOAD3	if A:=0 go to LOAD3;
402B	20 8B 40	LOAD2	JSR	PTRBYT	get data;
402E	91 FA		STA	FA,Y	store data;
4030	20 91 1F		JSR	\$1F91	compute checksum;
4033	20 63 1F		JSR	\$1F63	get next address;
4036	CA		DEX		X:=X-1;
4037	D0 F2		BNE	LOAD2	go to LOAD2;
4039	E8		INX		X:=X+1;
403A	20 8B 40	LOAD3	JSR	PTRBYT	get data;
403D	C5 F6		CMP	\$F6	compare high order checksum;
403F	D0 12		BNE	LOADER	if different go to LOADER;
4041	20 8B 40		JSR	PTRBYT	else get data;
4044	C5 F7		CMP	\$F7	compare checksum;
4046	D0 0B		BNE	LOADER	if different go to LOADER;
4048	8A		TXA		else A:=X;
4049	D0 B9		BNE	LOAD	if A not equal to 0 go to LOAD;
404B	A2 0C	LOAD7	LDX	0C	else X:= location of 'KIM';
404D	20 31 1E	LOAD8	JSR	\$1E31	output message;
4050	4C 4F 1C		JMP	START	go to START;
4053	A2 11	LOADER	LDX	11	X:=location of 'ERR KIM';
4055	D0 F6		BNE	LOAD8	go to LOAD8;
4057	A9 01	PTRINI	LDA	\$01	[initialization routine]
4059	8D 03 17		STA	\$1703	A:=B port address;
405C	8D 02 17		STA	\$1702	read flag:=1;
405F	AD 02 17		LDA	\$1702	A:=B register;
4062	29 02		AND	\$02	determine PB1;
4064	D0 08		BNE	OK	if reader ready go to OK;
4066	A9 58		LDA	'X'	else A:='X';
4068	20 A0 1E		JSR	\$1EAO	output 'X';
406B	4C 57 40		JMP	PTRINI	go to PTRINI;
406E	60	OK	RTS		return;
406F	AD 02 17	GETPTR	LDA	1702	[subroutine to input one character]
4072	29 20		AND	\$20	get bit from B data register;
4074	F0 F9		BEQ	GETPTR	if not ready go to GETPTR;
4076	A9 00		LDA	\$00	} else output read pulse;
4078	8D 02 17		STA	\$1702	
407B	A9 01		LDA	\$01	} turn off read pulse;
407D	8D 02 17		STA	\$1702	
4080	AD 02 17	CHECK	LDA	\$1702	} get bit 5 from B data register;
4083	29 10		AND	\$10	
4085	F0 F9		BEQ	CHECK	if character not ready go to CHECK;
4087	AD 00 17		LDA	\$1700	else get character;
408A	60		RTS		return;
408B	20 6F 40	PTRBYT	JSR	GETPTR	get character;
408E	20 AC 1F		JSR	\$1FAC	pack character;
4091	20 6F 40		JSR	GETPTR	get another character;
4094	20 AC 1F		JSR	\$1FAC	pack character;
4097	A5 F8		LDA	\$F8	A:=2 characters;
4099	60		RTS		return;
409A	20 57 40	MAIN	JSR	PTRINI	go to PTRINI;
409D	20 6F 40	LOOP	JSR	GETPTR	go to GETPTR;
40A0	4C 9D 40		JMP	LOOP	go to LOOP;
			END		

CROSS REFERENCE TABLE

Symbol	Value	Referenced
CHECK	4080	4085
GETPTR	406F	4004 4074 408B 4091 409D
LOAD	4004	4009 4049
LOADER	4053	403F 4046
LOADS	400B	****
LOAD2	402B	4037
LOAD3	403A	4029
LOAD7	404B	****
LOAD8	404D	4055
LOOP	409D	40A0
MAIN	409A	****
OK	406E	4064
PTRBYT	408B	4011 4018 4020 402B 403A 4041
PTRINI	4057	4001 406B 409A
PTRLD	4001	****
START	1C4F	4050

read until a semicolon is found; the next two characters give the hexadecimal number of bytes on the current line to be punched. This is followed by four characters, two bytes, giving the high order and low order bytes of the starting address for the data to follow. This is followed by the data which KIM-1 software always punches 24 bytes per line, a 2 character checksum for the line, and a carriage return. The carriage return is followed by six null characters, and a semicolon starts the next line. The last line punched contains 0 for the number of bytes, 0 for the address bytes and is followed by a four character checksum. When finished reading a paper tape, KIM-1 types 'ERR KIM' if the checksum does not compute (there has been an error in reading the tape), or just 'KIM' if the tape was read correctly.

The software consists of a copy of the KIM-1 monitor routine for reading paper tape modified by removing all calls to the GETCH and GETBYT routines and substituting two new routines, GETPTR and PTRBYT. A new subroutine, PTRINI is called at the beginning of the mainline program to properly configure the IO lines and check that the read head on the Fly Reader is closed. If it is not, KIM-1 will type out the character 'X' endlessly until the read head is closed. The software shown in listing 1 occupies 154 bytes starting at hexadecimal location 4000.

The Fly Reader is an excellent way to add a high speed paper tape reader to a microprocessor system. It is easy to interface and requires only a single +5 V supply. As a fast paper tape device it is considerably faster than an audio tape cassette system and offers increased flexibility of operation. ■

Listing 1: The basic software needed to run the Fly Reader with the KIM-1 microprocessor. The software uses the KIM-1 monitor routines and substitutes the GETCH and GETBYT routines with routines GETPTR and PTRBYT. The new subroutine PTRINI properly configures the IO lines used with the reader. Subroutine PTRINI will output an endless number of 'X' characters until the tape is loaded into the reader. The listing was set prepared from a cross-assembly provided by the author. A symbol table showing the values of the symbols used and where they are referenced follows the assembly and will prove useful when it is necessary to relocate a program at a different starting address.

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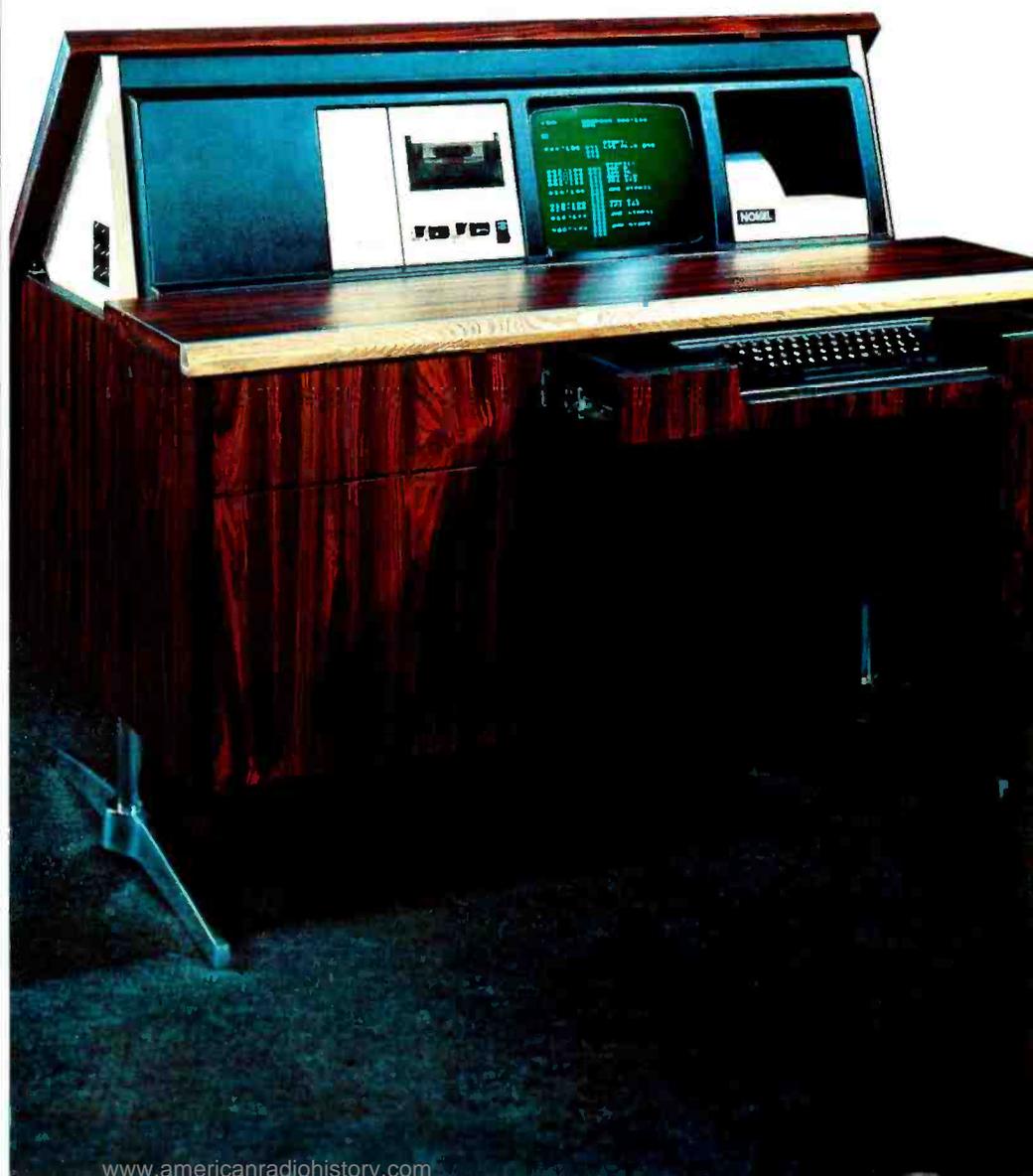


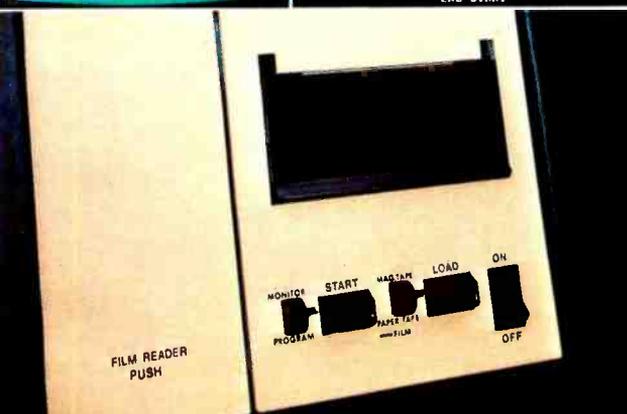
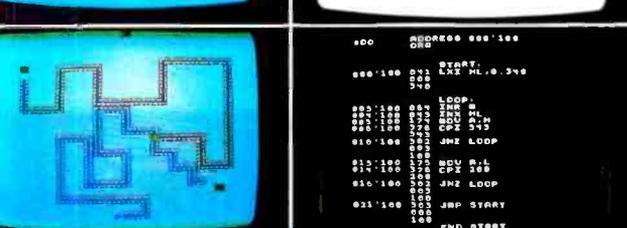
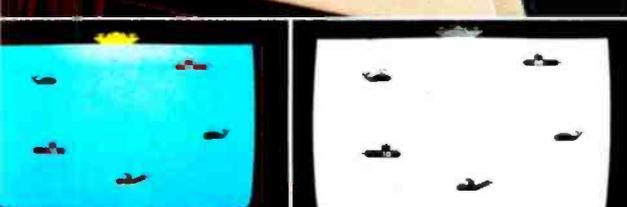
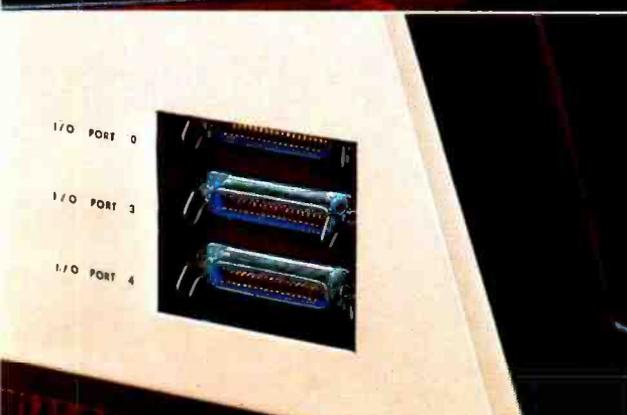
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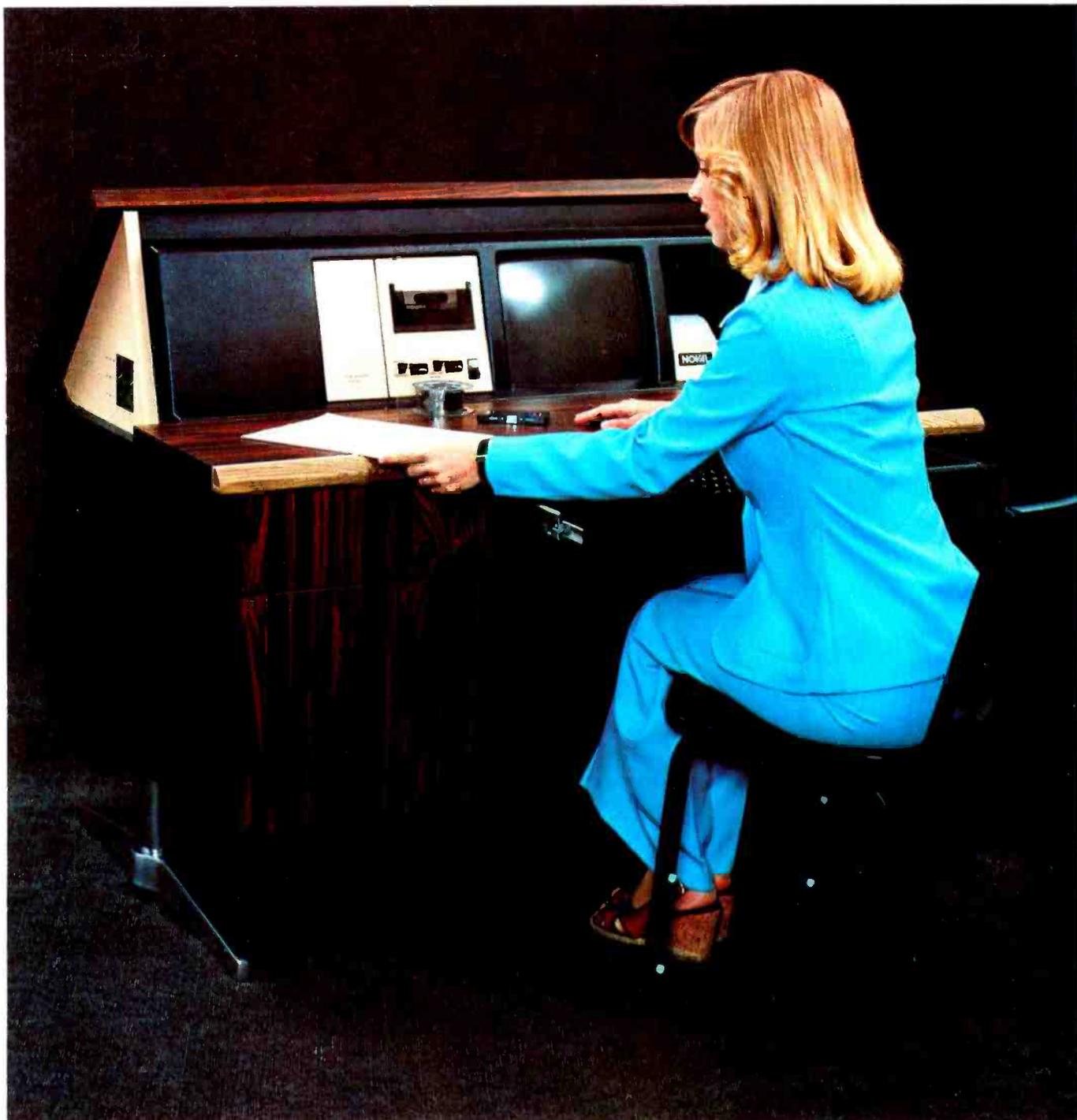
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FOR SALE: *Computers in Laboratory Medicine*. I have several copies of this book which are surplus to a class requirement. This excellent book is edited by Derek Enlander MD and is one of the most up to date and forward looking works on the subject of computerization in nuclear medicine, clinical pathology, anatomic pathology and diagnostic data retrieval. The book was published by Academic Press in 1975-1976 and costs \$14. David Johnson, 1473 Pine St, San Francisco CA 94109.

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FOR SALE: Friden Flexwriter Model 2303 with paper tape punch reader, 7 level, types 100 wpm in upper and lower case, works good, recently serviced, can be converted to ASCII, \$150 plus shipping. J E Upchurch, POB 1987, Sebring FL 33870. (813) 385-2788.

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FOR SALE: Altair 8800 computer with 2 MITS 4 K Dynamic memory cards, 88-ACR cassette interface, 88-2SIO serial board, 88-PIO parallel board, 2 expander boards, 8 K BASIC on cassette. Not completely assembled. Original cost was over \$1200. Will sell for \$750. Bob Majdanski, 214 Coolidge Av, Hasbrouck Heights NJ 07604. (201) 288-3742 after 7 PM.

WANTED: BYTE numbers 1, 2, 3, 8, 9 and 10. I will consider any offer, even partial ones. (One issue is better than none.) Send your terms and asking price to me and I'll get in touch with you. Tony Aiuto, 19 Old Field Ln, Lake Success NY 11020.

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FOR SALE: M6800 cross assembler. A two pass cross assembler written in FORTRAN IV is available for the M6800 Motorola micro. Input is in fixed format. Statements are similar to Motorola assembler language, most features of the language being supported. Additionally, a system symbol table is supported, enabling symbolic reference to system addresses and assembly of routines to contiguous memory locations. Send \$1 for the manual, and \$5 for the listing or \$8 for a paper tape (state if XOFF needed) to G A R Trollope, 433 Cherry Ln, Lewiston NY 14092. (716) 754-7222.

FOR SALE: CDP-1802 microcomputer software, Morse keyboard, 256 bytes, \$3. WAGUYV, 4956 Andrea Blvd, Sacramento CA 95842.

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WANTED: Someone who has or wants to write an 8 K version of BASIC for my COSMAC ELF, as in the *Popular Electronics* August 1976 construction article. It uses the RCA CDP 1802 central processor unit. Please send any and all correspondence on this subject to Greg Watson, 9617 Jomar Dr, Fairfax VA 22030.

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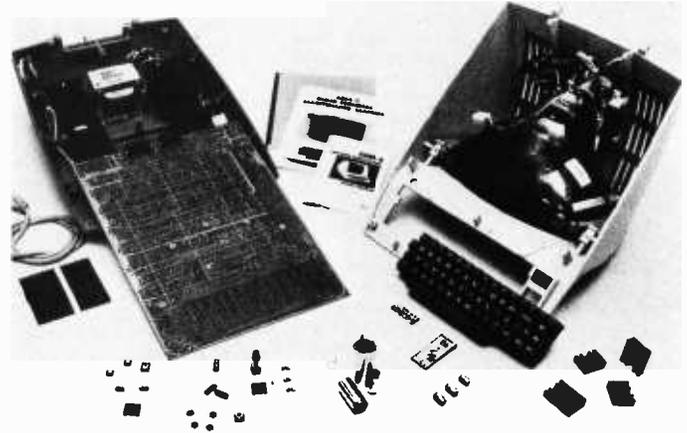
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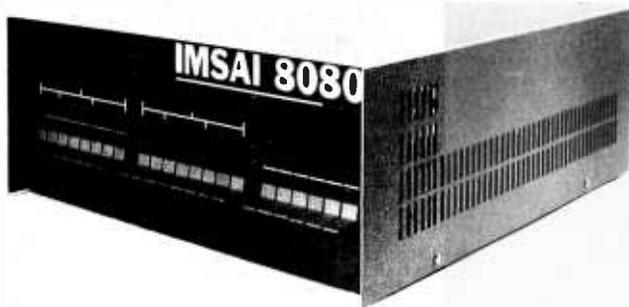
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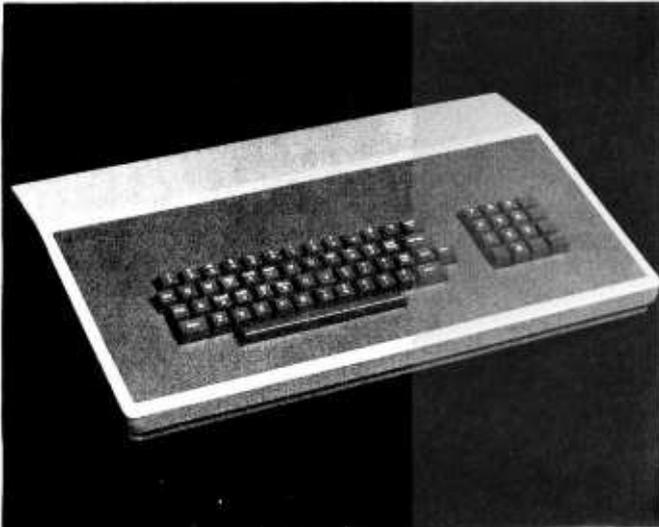
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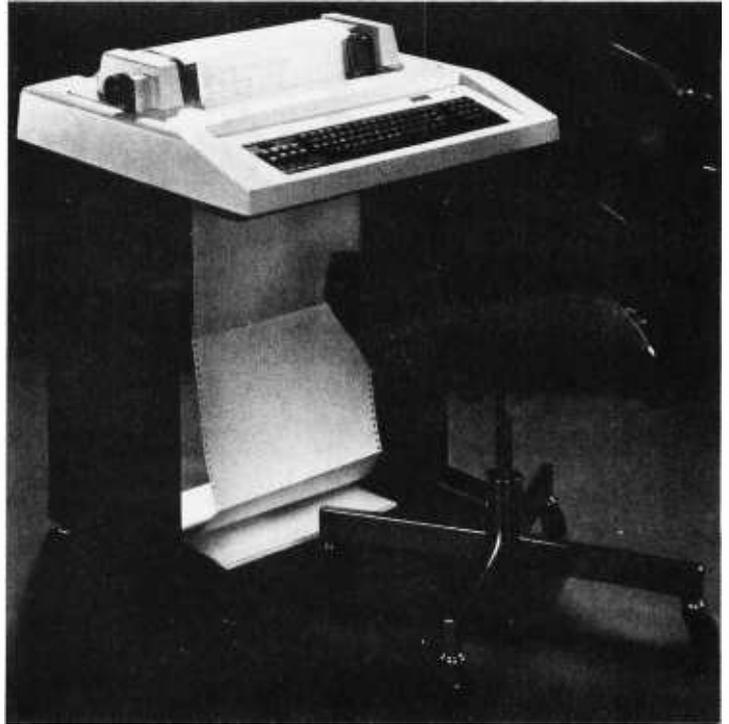
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Software for the Economy Floppy Disk

Dr Kenneth B Welles
General Electric, Nela Park
2623 Fenwick Rd
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The two fundamental routines needed for a floppy disk system are: Write a block of data to the disk, and read a block of data from the disk.

As hobbyists are rapidly finding out, even the most sophisticated hardware is next to useless without the proper software to control it. My previous article on the floppy disk drive interface (see February 1977 BYTE, page 34) described a hardware device of the simplest and, consequently, most software dependent type. This month's article describes the operation and use of the routines needed for transferring data between the computer and a disk drive (one of up to eight) connected to the interface.

The two fundamental routines needed for a floppy disk drive system are: Write a block of data to the disk, and read a block of data from the disk. This sounds simple in theory but in practice much more information and many operations are needed. How many bytes of data are in the block to be transferred, and which disk drive should the block go to or come from? At which track and sector is the block to be located? How will an error be detected (and if detected what steps should be taken for correction)? These are some of the major questions involved, without even considering such specific details as data format, file structure, unrecoverable errors, directory structures, naming and dating conventions, and so on ad futilitum.

A block read or write routine can be divided into four stages:

- Set up for the data transfer
- Transfer the actual data to or from the disk
- Error detection
- Error correction

These stages must, by their definition, occur in the order listed, and in most disk systems all four stages are included. In some operating systems the error detection and correction stages are ignored during a write operation. While this speeds up transfers by eliminating a reread or verify operation, it means that most write errors will be unrecoverable.

Write Set Up

Because this interface is quite unsophisticated, data to be written onto the disk must be prepared in memory in exactly the manner that it will appear on the disk. A preamble containing 16 bytes of zeroes (128 "0" bits) and a byte boundary synchronization signal (or sync byte) must precede the data. Before the data is written, the error detection bytes must be calculated and stored with the data to be written, since there will be no time to calculate them once the write operation has been initiated. This software uses a 16 bit cyclic redundancy check (CRC) word calculated from the data bytes by the binary polynomial:

$$x^{16} + x^{12} + x^5 + 1.$$

Because they are precalculated, the error detection bytes may be put in any position before, within, or after the data. I chose to place them directly after the data. The assembled block of data to be transferred (consisting of the preamble, sync byte, data and CRC bytes) is now in the proper format for transmission to the disk, but the disk drive itself is not yet properly set up for the

transfer. Because the interface may control multiple drives, the first action of the software is to select the proper drive and to assure that it is ready. "Ready" means that a diskette is loaded and revolving, and the power supplies are working. Next, the current track location of the data transfer head is determined and compared to the desired track. If the desired and current tracks differ, the software must step the head in or out at the proper rate (10 ms per track) until the correct track is reached, and then delay for the proper head settling time (10 ms) before continuing.

If the head is not currently loaded, the software must load the head and allow time for the loading to be accomplished (30 ms).

All that remains before initiation of the data transfer is to find the starting point of the proper sector. Because the timing is fairly critical between finding the sector and initiating the write operation, all of the parameters for the write data loop such as pointers, counters and output commands must be set up ahead of time in the 8080 registers. The software now searches for the index pulse from the index hole of the selected disk drive and, having found it, begins to count sector pulses until the desired sector is found. When the leading edge of this sector pulse is sensed, the software transfers immediately to the write routines, using the values previously stored in the registers to speed the initiation of the write operation. This generates the proper timing relationship between the sector pulse and the start of the recorded data.

Write Data Transfer

If N bytes of information (including the cyclic redundancy check) are to be recorded on the disk, the software will actually send out N+33 bytes of data to the interface. Figure 1 shows that the first 16 bytes (a preamble of zero bytes) are recorded to allow the interface to correctly distinguish between data pulses and clock pulses when this data block is read. The seventeenth byte recorded is the sync byte, in this case a binary value of 10000001. This sync byte is used by the Universal Synchronous Receiver Transmitter (USRT) to find the boundaries between bytes during a read operation. The next N bytes are the block of data to be recorded, and finally there are 16 bytes as a postscript to assure that no data is destroyed when the disk drive write gate is turned off. Because the only use of this data is to maintain clock synchronization and protect the preceding data, the contents of the last 16 bytes are not critical, and may overlap data used for some other purpose.

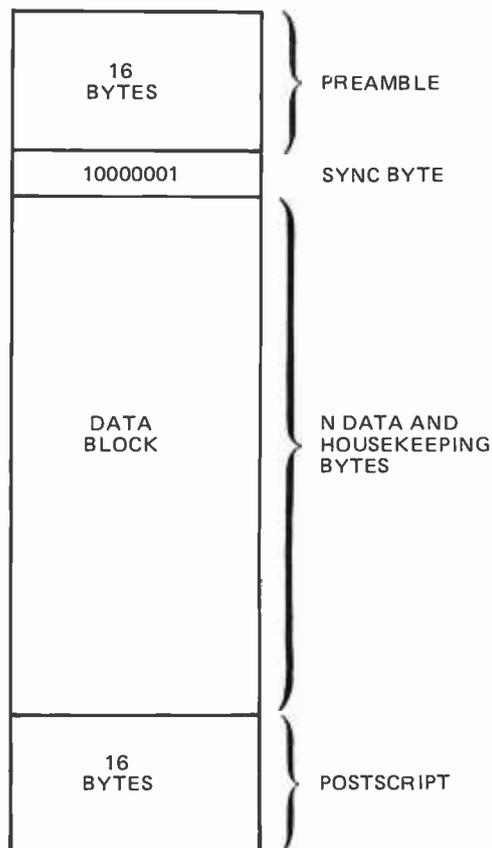


Figure 1: Floppy disk data transfer block format. The simplicity of this "hard sectored" floppy disk system requires that a specific format be used for the data. The first 16 bits act as a preamble to enable the interface to distinguish between data pulses and clock pulses. Next comes the sync byte used by the Universal Synchronous Receiver Transmitter (USRT) to find the boundaries between bytes. This is followed by the actual data and, finally, a 16 byte safety region to assure that no data is destroyed when the disk drive write head is turned off. See also table 2.

Write Error Detection

After all N+33 bytes have been sent to the disk through the interface, the write gate of the disk drive is turned off and error detection may now be performed. Error detection consists of performing a single read operation of the block of data just written. The block of data read in is compared byte by byte to the block of data written. If all N bytes are not the same for both blocks, an error has occurred. This process could be called a "verify" operation.

Write Error Correction

If error correction is necessary on a write operation that fails to verify, it is accomplished by rewriting the same block of data to the same disk, track and sector. After this rewrite attempt, a verify of the rewritten data is performed and compared to the correct data to determine the success or failure of the error correction operation. If

As block sizes increase there are fewer preambles and postscripts on any given track, thus maximizing the usable data bytes per track.

In the very simplest system each block of data would have some form of error detection ranging from a single byte of checksum to a 16 bit cyclic redundancy code or even a complex Hamming code.

the write operation fails four (or some other small integer) times in a row, all hope is abandoned, an error message is printed on the user's console device, and the write routine terminates. Manufacturers' recommendations for action taken at this point are as follows: Move to an unused sector and track on the disk in question, and retry the write operation. If the error persists, the disk drive has write circuitry problems; notify the user. If the error is eliminated, then the track and sector where the original error occurred probably has some damage to its oxide coating. In this case, relink the data file to reflect the new location of the data (the track and sector where the second attempt was made), and then record in some table the fact that the original track and sector where the write operation failed is an unusable area. But what do you do when the area that failed to write properly is the location of the table where the failed areas are stored? This is why operating systems designers have such a high incidence of insanity. The software in this article ignores the whole issue, the traditional ostrich solution.

Read Set Up

No data preparation need be done on a read operation because the only routines required are disk drive select, head load, track seek and sector seek. These are the same routines used by the write operation as described earlier.

Read Data Transfer

The USRT is reset to search for an occurrence of the sync byte within the incoming serial data, and the program is set to wait for the USRT to signal that the sync byte is found. When the sync byte is found, data is brought in from the interface and stored in memory. For a block recorded with N bytes of information as in the write operation, a total of N+1 bytes are read in. The first byte brought in by the software is the sync byte, accounting for the extra byte transferred.

Table 1: Characteristics of different data block sizes. Smaller data blocks have the advantage of not tying up large blocks of memory (a premium commodity in many small systems). Large data blocks, by comparison, speed up data transfer, require fewer blocks per track, and maximize the number of usable data bytes per track.

Data Blocks Per Track	Usable Bytes Per Block	Data Bytes Per Block	Housekeeping Bytes Per Block
1	5127	4096	1031
2	2549	2048	501
4	1260	1024	236
8	615	512	103
16	293	256	37
32	132	128	4

All Formats Store 315,392 Data Bytes Per Disk

Read Error Detection

The routine that calculated the 16 bit cyclic redundancy check (CRC) for the write operation is now used to calculate the CRC of the data block just read. The calculated cyclic redundancy check must match the CRC read in if the data is correct. If these values do not agree, an error has occurred during the read operation.

Read Error Correction

If a single read operation is unsuccessful, two more attempts are made to read the data. If the computed and read-in cyclic redundancy check values still disagree, the disk head is stepped in one track and then out one track, and up to three more read operations are tried. Continued lack of success causes the head to be stepped out and then stepped in one track, and three more attempts to read are made. If the error persists after all of this, then this software concludes that the data is unrecoverable. The jogging of the head one track in or out is recommended by the manufacturer to release any dust particles which may have lodged between the disk and head, and which therefore may be causing the read errors.

Block Size

The discussion of the software to this point has referred to the transfer of a "block" of data, but no definition has been made of the size of this block. The disk is divided physically into 32 different sectors by the presence of 32 sector holes. The combination of the rotational speed of the disk (360 rpm), the data rate (250,000 bps), variation in the timing of the sector hole detection ($\pm 500 \mu s$), and the necessity for the 16 byte preamble, sync byte and 16 byte postscript on each record determines that if each track holds 32 blocks of data, then each block has 132 usable data bytes. There is no electronic or philosophical reason that the disk must be recorded in a format of 32 data blocks per track, or even that a 32 block track must consist of full size 132 byte blocks. The proper operating system software could quite easily make use of a data format with only 16 bytes per block and 11 blocks per track. However, this would make the capacity of each disk less than 30 K bytes, a tenfold reduction in storage capacity. The software presented here can be easily modified to allow other block sizes and formats to be used, and the reasons for choosing different block sizes will be presented before attempting to justify the block size used.

Smaller data blocks have three main advantages. First, because of the nature of a disk drive, data must be transferred one block at a time, never as a fraction of a block. If the data is not ready all at once (as is usually the case), then some area of computer memory must be dedicated to the storage of this data until a full block can be acquired. The size of this buffer area is the same as the size of the block, and a smaller data block consumes less memory than a large data block. Even with the dropping prices of memory, the average person cannot afford to dedicate large blocks of a computer's storage solely to disk data buffering, especially in a sophisticated operating system that may work on many disk data files simultaneously and require a data buffer for each one. Secondly, when recording data files onto a disk device, the data file rarely contains the right number of bytes to be stored in exactly an integral number of blocks. An average of one-half of a data block is wasted on each different file recorded on the disk. If one works with a data base that consists of a large number of short data files, this lost capacity of this disk can become significant. Smaller block size minimizes this loss of storage capacity. Last and definitely not least, the growing awareness of the need for standardization of data storage among the users is a motivation to seek out de facto (read IBM) standards to work with. Operating with somebody else's block and physical recording parameters makes it potentially possible to exchange data between different computer systems via diskettes. The most widely embraced standard for floppy disk data operations is the IBM soft sectored standard. Unfortunately, this standard is incompatible with the inexpensive interface I described in my previous article, and does not (at the time of this writing) lend itself to as cheap and simple an interface. However, the IBM standard uses 128 byte data blocks, and large amounts of software exist for it. Software for the hard sectored version of this format currently uses data blocks of 128 byte size, although the new double density disk drives are promoting software with 256 byte data blocks. Most existing software, then, is written with 128 byte blocks of data, 32 blocks per track on a hard sectored disk or 26 blocks per track on a soft sectored disk. This is an argument in favor of short data blocks, based on conventions of existing users of floppies.

Larger block sizes also have three main advantages. First, because the software presented here can only transfer one block of data per revolution, large data blocks greatly

speed up data transfer. Dedicating a track to a single data block, a 5 K byte block of data can be transferred in 1/6 second. Second, large block sizes mean fewer blocks per disk. This reduces the data required in the addressing of disk data blocks, and can reduce the complexity of directories and block occupancy bit maps. Table 1 shows the third advantage of large data block sizes. As block sizes increase there is less overhead on any given track, and so the maximum number of usable data bytes per track (and per disk) increases. While the overall amount of data storage increases by only a small percentage, the increase is important to certain data structures, as will be shown.

In block structured random access data bases, such as disk storage, data is usually stored in blocks of 2^n bytes. An integral power of two is a handy size for organization of data in a binary computer and makes data manipulation easier than with some other sizes. With each block of useful data, there is some other data which is transparent to the user and of no real concern to him/her, but which is associated with file management and the operating system of the computer. This housekeeping or "overhead" data contains information about the status of the data within the block. In the very simplest system, for instance, each block of data would have some form of error detection, ranging from a single byte of checksum to a 16 bit cyclic redundancy code or even a complex error correcting Hamming code. Most simple systems also record the track number and sector number of the block with

No other peripheral device requiring direct memory access (DMA) operations should be active during the read or write operations of this software. In general, no interrupts or hold operations lasting longer than two microseconds can occur simultaneously with disk writing.

Table 2: This is an expansion of figure 1 showing how the data block is laid out in memory prior to an output disk transfer or following a read operation from the disk. The "housekeeping" bytes mentioned in the text are used for error detection and error correction as well as for labeling files.

Byte Number (Decimal)	Value	Function
-17 to -2	0	Clock synchronization
-1	hexadecimal 81	Sync byte for data synchronization
0 to 8	ASCII	9 letter file name
9	0 to 7	Drive number
10	3 to 76	Track number
11	0 to 30	Sector number (must be even)
12	3 to 76	Track number of preceeding block
13	0 to 30	Sector number of preceeding block
14	3 to 76	Track number of following block
15	0 to 30	Sector number of following block
16 to 19	-	Future use
20	0 to 255	Byte count of incomplete data blocks
21 to 276	DATA	User data (256 bytes)
277	-	Future use
278 to 279	CRC	16 bit CRC (cyclic redundancy check)
280 to 295	0	Trailer bytes

Bytes 0 to 279 are shown as they exist in memory.

Bytes -17 to 295 are shown as they exist on the disk.

NOTE: The software in this article uses only bytes 9 to 11 and 278, 279. (All bytes numbered 0 to 279 are transferred to disk and back.) See note at end of article concerning a complete floppy disk operating system which uses all the bytes.

```

; THE FOLLOWING IS A GENERAL PURPOSE ERROR TYPEOUT ROUTINE
; THE VALUE OF CO REFERS TO A USERS ROUTINE WHICH TYPES OUT
; THE ASCII VALUE IN THE C REGISTER ONTO THE USERS CONSOLE OUTPUT.
; ONLY THE A REGISTER AND FLAGS MAY BE DESTROYED BY CO.

0000 CO EQU 0
0010 ORG 10H ; THIS ROUTINE MAY BE ORG'D ANYWHERE IN MEMORY

; THIS ROUTINE TYPES OUT AN ERROR MESSAGE TO THE USER,
; IT IS CALLED INTERNALLY AS:
; CALL ERTYP
; DB X
; THE RESULT IS THE TYPING OF:
; ERROR X

0010 E3 ERTYP: XTHL
0011 F5 PUSH PSW
0012 C5 PUSH B
0013 D5 PUSH D
0014 7E MOV A,M
0015 23 INX H
0016 E5 PUSH H
0017 F5 PUSH PSW
0018 212000 LXI H,ERMES
001B CD5500 CALL TXTYP
001E F1 POP PSW
001F CD2F00 CALL BYTYP
0022 E1 POP H
0023 D1 POP D
0024 C1 POP B
0025 F1 POP PSW
0026 E3 XTHL
0027 C9 RET

0028 0D0A4552 ERMES: DB 13,10,'ERPO','R'+128
002C 524FD2

002F F5 BYTYP: PUSH PSW
0030 0E20 MVI C,' '
0032 CD0000 CALL CO
0035 F1 POP PSW
0036 F5 NMTYP: PUSH PSW
0037 0F RRC
0038 0F RRC
0039 0F RRC
003A 0F RRC
003B CD4300 CALL HEXCHR
003E F1 POP PSW
003F CD4300 CALL HEXCHR
0042 C9 RET

0043 F5 HEXCHR: PUSH PSW
0044 E60F ANI 0FH
0046 C630 ADI '0'
0048 FE3A CPI '9'+1
004A DA4F00 JC HEX1
004D C607 ADI 'A'-'9'-1
004F 4F HEX1: MOV C,A
0050 CD0000 CALL CO
0053 F1 POP PSW
0054 C9 RET

0055 7E TXTYP: MOV A,M
0056 B7 ORA A
0057 C8 RZ
0058 F5 PUSH PSW
0059 E67F ANI 7FH
005B 4F MOV C,A
005C CD0000 CALL CO
005F F1 POP PSW
0060 F8 RM
0061 23 INX H
0062 C35500 JMP TXTYP

0001 ERWRT EQU 1 ; IF THE DISK FAILS TO WRITE, TYPE "ERROR 1"
0002 ERRED EQU 2 ; IF THE DISK FAILS TO READ, TYPE "ERROR 2"
0003 ERDKNM EQU 3 ; IF NO SUCH DISK EXISTS, TYPE "ERROR 3"

; THESE ARE THE FLOPPY DRIVE I-O ROUTINES
; 19-AUG-76

00F0 INPORT EQU 0F0H
00F0 OTPORT EQU 0F0H

00F0 FDRED EQU INPORT ; DATA FROM THE FLOPPY
00F1 FDSTAT EQU INPORT+1 ; STATUS BITS OF THE FLOPPY
00F2 SRTSWE EQU INPORT+2 ; READ THE SRT STATUS WORD
00F3 SRTRP EQU INPORT+3 ; INPUTTING THIS RESETS THE USRT TO LOOK
00F4 FDIWNT EQU INPORT+4 ; KLUGE TO SYNCHRONIZE TO 32 MICROSECS

00F0 FDWRIT EQU OTPORT ; DATA TO BE WRITTEN TO THE FLOPPY
00F1 SRTFS EQU OTPORT+1 ; PORT OF THE DEFAULT CHARACTER
00F2 SRTRSS EQU OTPORT+2
00F3 FDOUIT EQU OTPORT+3 ; SEND SIGNALS TO THE FLOPPY
00F4 FDOTMT EQU OTPORT+4 ; SYNC TO 32 MICROSECS
00F5 LOADHD EQU OTPORT+5 ; THIS COMMANDS LOADS THE HEAD FOR 3 SECONDS
00F6 HDUNLD EQU OTPORT+6 ; THIS FORCES AN IMMEDIATE HEAD UNLOAD

; THESE ARE SIGNALS TO FDOUIT (DIRECT FLOPPY CONTROLS)
0001 WCSEL EQU 1 ; 1=TRACK 0-43, 0=TRACK 44-76
0002 FUREST EQU 2 ; USE TO RESET A FILE UNSAFE CONDITION
0004 DIRIN EQU 4 ; LOGICAL OR FOR IN TRACK
FFFB DIROT EQU NOT DIRIN ; AND FOR OUT TRACK
0009 WRTGAT EQU 100 ; LOGICAL OR TO START A WRITE
0010 STEPP EQU 200 ; LOGICAL OR TO START A STEP PULSE
FFEF STEPM EQU NOT STEPP ; LOGICAL AND TO STOP THE PULSE

; THESE ARE THE DIRECT STATUS LINES OF THE FLOPPY (IN FDSTAT)
0001 TRKZRO EQU 1
0002 UNSAFE EQU 2
0004 FDRDY EQU 4

```

Listing 1: Floppy disk drive input and output routines. As it stands, this listing is one step short of being a complete operating system, but could be used as a starting point for such an operating system. This program will run only on an 8080 with memory having a cycle time of 500 ns or faster; it will not run correctly if any interrupts or direct memory access (DMA) routines lasting longer than 2 μ s occur during program execution. (See part 1, page 42, February 1977 BYTE, "Software Timing.")

the rest of the data on that block. When some track and sector is read, the track and sector numbers read must agree with the desired track and sector numbers. If these bytes disagree, then a seek error has occurred. The error detection information and the track and sector numbers are simple housekeeping data that the user need never be concerned with during data transfer. A more complex data base management system may store extensive housekeeping information with each block of data. Information such as forward and reverse linkages (pointers to the succeeding and preceding blocks of data in a file), file name, number of valid bytes in a partially filled data block, date and time of the recording of this block of data, write or read protection, user identification, destroy date — all may be recorded as useful housekeeping data.

If a format of 32 blocks per track, 128 data bytes per block is used, table 1 shows that this leaves only four bytes of data available with each block to record housekeeping data. Going to 16 blocks of 256 data bytes per block, the available housekeeping area increases to 37 bytes per block. 512 data byte blocks gives a substantial 103 bytes of housekeeping on each of eight blocks per track. The routines presented here were written as a first low level step toward a full floppy disk operating system, and the amount of housekeeping required for the full system necessitated at least the 256 byte block size. The 512 byte block size was ruled out because it consumed buffer space much too quickly, and because a single byte would no longer be sufficient as a buffer data counter.

A DOS Block Format

Table 2 shows the layout of the data block for my disk operating system (DOS) as it exists in memory and as it is written to and read from the disk. There are 280 bytes in the block, 256 of which are usable data bytes, and 24 of which are reserved for housekeeping. Of the housekeeping bytes,

five are not currently defined, and all but bytes 9, 10, 11, 278 and 279 are available to the user in the routines listed in this article. As of this writing, a full floppy disk operating system for my peripheral interface to the 8080 has been completed (see note at end of this article). The format described here is that used by this operating system. In this format, the first nine bytes are designated as the file name, ordered a la DEC as six ASCII characters of file name and three characters of extension. Having the file name associated with each block of a file is a great aid in data recovery from crashed systems (a not unheard of occurrence) and in troubleshooting modifications to the operating system (a frequently heard of occurrence). Byte 9 is the number of the device (disk 0 to disk 7) where this file was originally recorded. This number does not have to match the number of the device that the file is subsequently read from. Bytes 10 and 11 are the numbers of the track and sector where this data block is stored. Valid values are any integer from 0 to 76 for the track, and only even integers from 0 to 30 for the sector. (With 256 byte blocks, we have to use every other sector.) Bytes 12 and 13 indicate the track and sector of the block of data preceding this one in a linked data file. If the current block is the first block, and no previous data blocks exist, then bytes 12 and 13 are set to zero. Since track zero, sector zero is reserved for the bootstrap copy of the operating system, no linked file can have this block address as a legal forward or reverse linkage. Bytes 14 and 15 give the track and sector of the data block following this block in the linked file. If this is the last block, values of 0 and 0 are again used to indicate an invalid linkage (end of file).

Because no operating system is ever finished and better ideas are generated as a continuous function of time, bytes 16 to 19 are reserved for whatever functions may become desirable in the future. Byte 20 is used when the last block in a recorded data file is not completely filled, indicating the actual number of valid data bytes stored in the data area. Bytes 21 to 276 inclusive contain the 256 bytes of usable data. The data in these bytes is usually the only data of final concern to a program retrieving data from a file. Bytes 278 and 279 are the cyclic redundancy check bytes, least significant byte first. This 16 bit value is calculated as detailed earlier, from the values of bytes 0 to 276 inclusive.

Now, the Software

Listing 1 shows the software which controls the disk interface and performs the

Listing 1, continued:

```

0008      SCTR EQU      8
0010      INDX EQU     10H
0020      HEADLD EQU   20H

; THESE ARE VARIOUS PARAMETERS
0001      SYNCB EQU     01H ; THIS IS THE SYNC BYTE FOR START OF TRACK
0100      LNBUF EQU     256 ; LENGTH OF THE BUFFER
0016      XTRA EQU     22 ; EXTRA BYTES FOR HOUSEKEEPING
0117      LN2SNC EQU   LNBUF+XTRA+1 ; DISTANCE OF SYNC TO CHECKSUM
0008      DSKLIM EQU    8 ; UP TO 8 DISK DRIVES ON THIS CONTROLLER

E000      ORG      0E000H ; THIS IS DISCONTIGUOUS RAM

E000 F3      GDWRT: DI ; DISABLE THE INTERRUPT TO MAINTAIN CRITICAL TIMING
E001 0602     MVI   B,2 ; TRY THIS 3 TIMES BEFORE GIVING UP
E003 C5      GDWR1: PUSH B
E004 CD15E0   CALL  WRT256 ; FIRST, WRITE IT
E007 CD6FE0   CALL  CHKWRT ; THEN CHECK FOR PROPER WRITE
E00A C1      POP   B ; RESTORE STACK, DON'T CHANGE FLAGS
E00B C8      PZ    ; SUCCESSFUL WRITE, PETPUN NOW
E00C 05      DCR   B ; A BAD WRITE, TRY AGAIN
E00D F203E0  JP    GDWR1 ; IF WE STILL HAVE TIME, TRY AGAIN
E010 CD1000  CALL  ERTYP ; ELSE, SIGNAL THE ERROR AND QUIT
E013 01      DB   ERWRT
E014 C9      RET

; THIS ROUTINE ASSUMES THAT THE TRACK AND SECTOR TO BE WRITTEN ARE IN TRKWRT
; AND THE DATA IS ALL IN WRDAT
E015 3A65E2  NPT256: LDA   WRDEV ; THIS MUST CONTAIN THE DISK NUMBER!!!
E018 CD28E1  CALL  DSKNUM ; PROCESS ANY UNIT NUMBER CHANGE
E01B 2148E2  LXI   H,WRWBUF ; CLEAR THE CLOCK SYNC BYTES
E01E AF      XRA   A ; BY FILLING WITH ZEROES
E01F 0610     MVI   B,10H ; 16 OF THEM
E021 77      CLRLOP: MOV  M,A
E022 23      INX   H
E023 05      DCR   B
E024 C221E0  JNZ   CLRLOP ; STORE THE SYNC BYTE
E027 3691     MVI   M,SYNCB
E029 2A66E2  LHLD  TRKWRT
E02C 2248E2  SHLD  TRKWRT ; SET UP THE TRACK AND SECTOR POINTS
E02F 2158E2  LXI   H,WRSNCB ; START OF THE CHECKSUM AREA
E032 CDF3E0  CALL  CHK436
E035 EB      XCHG ; STORE THE CALCULATED CHECKSUM
E036 2272E3  SHLD  WRCHKS ; GET THE DESIRED TRACK
E039 CD08E1  CALL  TRKGET ; LOAD THE HEAD
E03C CD09E2  CALL  HEDLOD ; BUFFER AREA TO WRITE TO THE DISK
E03F 113691  LXI   D,LNBUF+XTRA+32
E042 2148E2  LXI   H,WRWBUF
E045 AF      XRA   A
E046 D3F1     OUT  SRTTFS ; TRANSMIT A FILL CHAR OF ZERO
E048 3A46E2  LDA   FDBUF ; THE OUT STATUS
E04B F60A     ORI   WRTGAT+FUREST ; TURN ON THE WRITE GATE
E04D 4F      MOV  C,A ; HOLD THE WRITE COMMAND IN C
E04E CD1AE2  CALL  SCTGET ; GET THE PROPER SECTOR
E051 79      MOV  A,C
E052 D3F3     OUT  FDOUT ; INIT THE WRITE GATE QUICKLY
E054 E6FD     ANI  NOT FUREST ; TURN OFF FILE UNSAFE RESET
E056 D3F3     OUT  FDOUT
E058 D3F4     WRTLOP: OUT  FDOTWT ; KLUGE AND WAIT FOR READY FOR DATA
E05A 7E      MOV  A,M
E05B D3F0     OUT  FDWRIT
E05D 23      INX   H
E05E 1B      DCX   D
E05F 7A      MOV  A,D
E060 03      ORA   E
E061 C258E0  JNZ   WRTLOP
E064 3A46E2  LDA   FDBUF
E067 E6F7     ANI  NOT WRTGAT ; TURN OFF THE WRITE GATE
E069 D3F3     OUT  FDOUT ; NOW
E06B 3246E2  STA  FDBUF ; THIS IS THE PRESENT STATUS
E06E C9      RET

; THIS ROUTINE PERFORMS A REREAD OF THE SECTOR JUST WRITTEN.
; RETURN IS WITH Z=0 IF A READ FAIL OCCURRED, OR IF A READ DIDN'T MATCH
; THE IMMEDIATE PREVIOUS WRITE. Z=1 IS A SUCCESSFUL WRITE
E06F CDA3E0  CHKWRT: CALL  TRV3RD ; TRY 3 TIMES TO READ IT
E072 C0      RNZ ; RETURN WITH Z=0 ON A BAD ERROR
E073 2175E3  LXI   H,DRSNCB ; SET UP A 256 BYTE COMPARE
E076 1158E2  LXI   D,WRSNCB
E079 0600     MVI   B,0 ; 256 ISN'T ALL, BUT IT IS ENOUGH
E07B 1A      CHKW1: LDAX  M ; GET WHAT WAS WRITTEN
E07C BE      CMP  M ; COMPARE TO WHAT WAS READ
E07D C0      RNZ ; RETURN ON AN ERROR
E07E 13      INX   D
E07F 23      INX   H ; UP THE POINTERS
E080 05      DCR   B
E081 C27BE0  JNZ   CHKW1
E084 C9      RET ; SUCCESSFUL RETURN, Z=1

; THIS ROUTINE TRIES 3 TIMES TO READ, THEN JOGS IN, AND OUT AND AGAIN
RED256: DI ; DISABLE THE INTERRUPT FOR CRITICAL TIMING
E085 F3      DI ; TRY 3 TIMES TO READ THE BLOCK
E086 CDA3E0  CALL  TRV3RD ; IF SUCCESSFUL, THEN RETURN
E089 C8      RZ ; JOG IN
E08A CDA3E1  CALL  TRAKIN ; SCRAPE OFF FLIES AND FROGS
E08D CD02E1  CALL  TRAKOT ; THREE MORE
E089 CDA3E0  CALL  TRV3RD
E093 C8      RZ
E094 CD02E1  CALL  TRAKOT ; JOG OUT
E097 CDA3E1  CALL  TRAKIN ; SCRAPE OFF DIGITS AND DOGS
E09A CDA3E0  CALL  TRV3RD ; LAST CHANCE COMBOY
E09D C8      RZ ; IF FINALLY SUCCESSFUL, THEN RETURN
E09E CD1000  CALL  ERTYP ; IF NOT, THEN TYPE OUT THE ERROR
E0A1 02      DB   ERRED
E0A2 C9      RET

; THIS ROUTINE TRIES TO READ THREE TIMES, RETURNS Z=1 FLAG IF OK ON REA
E0A3 3E02     TRV3RD: MVI  A,2 ; ACTUALLY TRY 3 TIMES
E0A5 3245E2  STA  REDTRY ; HOLD ONTO IT
E0A8 CDB4E0  TRGP: CALL REDONC ; TRY IT JUST ONCE
E0AB C8      RZ

```

Listing 1, continued:

```

E0A0 2145E2      LXI   H,REDTPV
E0A1 35          DCR   M
E0B0 F8         RM    ; IF TOO MANY, THEN RETURN Z=0
E0B1 C3A8E0     JMP   TRGP ; TRY TRY AGAIN, AGAIN

; THIS ROUTINE TRIES ONLY ONCE (AND I DON'T BLAME IT ATALL)
E0B4 3A4AE2     PEDONC: LDA   DSKWNT ; THIS MUST CONTAIN THE DISK NUMBER!!
E0B7 CD2BE1     CALL  DSKNUM ; PROCESS ANY UNIT CHANGE
E0BA CD08E1     CALL  TRKGET ; GET THE TRACK
E0BD CD09E2     CALL  HEDL0D ; LOAD THE HEAD
E0C0 111A01     LXI   D,LNBUF+XTRA+4
E0C3 2175E3     LXI   H,RDSNCB ; POINT TO THE BUFFER AREA
E0C6 3E01     MVI   A,SYNCSB ; PROMPT THE USRT WITH WHAT TO EXPECT
E0C8 D3F2     OUT   SRTRSS ; SHOVE IT
E0CA CD1AE2     CALL  SCTGET ; GET THE PROPER SECTOR
E0CD DBF3     IN   SRTAR ; RESET THE USRT TO LOOK FOR THE SYNCSB
E0CF DBF4     REDLOP: IN   FDINWT ; SPECIAL SYNC KLUGE TO HOLD ONTO THE BU
E0D1 DBF0     IN   FDRED ; UNTIL DATA IS AVAILABLE
E0D3 77       MOV   M,A
E0D4 23       INX   H
E0D5 1B       DCX   D
E0D6 7A       MOV   A,D
E0D7 02       ORA   E
E0D8 C2CFE0     JNZ   REDLOP
E0DB 2175E3     LXI   H,RDSNCB ; POINT TO THE BUFFER AREA
E0DE CDF3E0     CALL  CHK43B
E0E1 3A3AE2     LDA   LSTDSK ; GET THE NUMBER OF THIS DISK,
E0E4 327FE3     STA   RDDEV ; AND STORE IT IN THE READ AREA FOR FUTURE USE
E0E7 2A9CE4     LHLD  RDCHKS ; HL HAS THE CHECKSUM
E0EA C3EDE0     JMP   COMPAP

; THIS ROUTINE COMPAPES HL TO DE
E0ED 7C       COMPAP: MOV   A,H
E0EE 92       SUB   D
E0EF C0       RNZ
E0F0 7D       MOV   A,L
E0F1 93       SUB   E
E0F2 C9       RET

; THIS ROUTINE DOES A DOUBLE PRECISE CHECKSUM
E0F3 011601     CHK43B: LXI   B,LNBUF+XTPA
E0F6 110000     CHKSUM: LXI   D,0
E0F9 7E       CHKLOP: MOV   A,M ; GET THE BYTE TO SUM
E0FA E5       PUSH  H
E0FB C5       PUSH  B ; SAVE ALL THE REQS.
E0FC AB       XRA   E
E0FD 47       MOV   B,A
E0FE 0F       RRC
E0FF 0F       RRC
E100 0F       RRC
E101 0F       RRC
E102 4F       MOV   C,A
E103 AB       XRA   B
E104 E6F0     ANI   0F0H
E106 AA       XRA   D
E107 6F       MOV   L,A
E108 79       MOV   A,C
E109 07       RLC
E10A E61F     ANI   1FH
E10C AD       XRA   L
E10D 6F       MOV   L,A
E10E 70       MOV   A,B
E10F 07       RLC
E110 E601     ANI   1
E112 AA       XRA   D
E113 AD       XRA   L
E114 57       MOV   D,A
E115 79       MOV   A,C
E116 E60F     ANI   0FH
E118 AB       XRA   B
E119 5F       MOV   E,A
E11A 79       MOV   A,C
E11B AB       XRA   B
E11C 07       RLC
E11D E6E0     ANI   0E0H
E11F AB       XRA   E
E120 5F       MOV   E,A
E121 C1       POP   B
E122 E1       POP   H
E123 23       INX   H
E124 0B       DCX   B
E125 76       MOV   A,B
E126 01       ORA   C
E127 C2F9E0     JNZ   CHKLOP ; NEXT BYTE IF NOT YET DONE
E12A C9       RET

; THIS ROUTINE IS ENTERED WITH A CONTAINING THE NUMBER OF THE
; DISK DRIVE UNIT ON WHICH THE DESIRED OPERATIONS ARE TO BE PERFORMED.
; (0-DSKLIM-1). IF THE LAST DRIVE USED WAS THIS DRIVE, THEN
; RETURN IS IMMEDIATE, AND NO ACTION IS TAKEN. IF A NEW DRIVE
; IS CALLED, THEN THE HEAD OF THE PRESENT DRIVE IS FORCED TO UNLOAD
; AND THE CURRENT TRACK NUMBER ASSOCIATED WITH IT IS STORED.
; THE NEW DISK'S LAST TRACK IS REMEMBERED, AND IF VALID, CONTROL IS
; RETURNED. ELSE, THE NEW DISK IS INITED, AND THE TRACK 0 IS FOUND.

E12B FE00     DSKNUM: CPI   DSKLIM ; 0-7 IS CONDITIONALLY ACCEPTABLE
E12D F285E1     JP    DSKR1 ; NO SUCH DISK EXISTS!!
E130 213AE2     LXI   H,LSTDSK ; POINTER TO THE LAST DISK
E133 0E       CMP   M ; NEW=OLD?
E134 C0       RZ ; IF SO, THEN RETURN NO ACTION
E135 F5       PUSH  PSW ; ELSE, SAVE THE NEEDED DATA
E136 E5       PUSH  H
E137 3E00     MVI   A,DSKLIM ; WAS THE OLD ONE LEGAL?
E139 0E       CMP   M
E13A FA46E1     JM    DSKN1 ; IF NOT, DON'T STORE THE OLD TRACK
E13D 4E       MOV   C,M ; C=LAST NUMBER
E13E 0600     MVI   B,0
E140 09       DAD   B
E141 23       INX   H ; HL=LSTDSK+OLD DISK+1

```

operations described above. There are three main entry points to this software: DKINT, GDWRT and RED256. When called, these routines initialize a disk drive, write a data block, and read a data block, respectively.

DKINT

DKINT (for DisK INiTialize) is a subroutine called with the accumulator containing the number (0 thru 7) of the disk drive to be initialized. DKINT causes the selected disk drive to unload its data transfer head, step the head to track zero, and reset the track counter associated with that drive to zero. DKINT destroys the contents of all of the 8080 registers, so appropriate precautions should be taken when using it. This routine must be used after powering up the processor, and should be used both before disks are removed from the disk drives, and before they are referenced by programs after insertion.

GDWRT

GDWRT (for Good WRiTe) causes a data block to be written to a specified track and sector on a specified disk. When the GDWRT subroutine is called, all of the data relevant to the transfer is assumed to already reside in the 280 byte buffer area from WRFLNM to WRFLNM+279. The number of the disk to be written on (0 thru 7) must be in location WRDEV (which is WRFLNM+9), the track number (0 thru 76) must be in TRKWRT (which is WRFLNM+10), and the sector number (an even number from 0 to 30) must be in TRKWRT+1 (WRFLNM+11). The data to be stored on the disk may reside anywhere in bytes 0 to 8 and 12 to 277 of the buffer area, but in the context of my operating system, the 256 bytes from WRFLNM+21 to WRFLNM+276 are specifically reserved for data. GDWRT performs the steps previously described for the write data block operation: generating the preamble, calculating the cyclic redundancy check, selecting the disk and positioning the head, writing the block, verifying the write operation and attempting to rewrite if necessary. Control will return to the program which called GDWRT after a delay of from 0.2 to 2 seconds, depending on the distance that the head had to travel and the occurrence of any write errors. The user need not worry about the timing of the operation; however, no interrupts or hold operations longer than 2 μ s may occur during GDWRT. Specifically, no memory slower than 500 ns response time may be used, and no device using direct memory access (DMA) operations may occur during GDWRT. If such a delay happens at the wrong time, the data

will not be properly recorded, and following data blocks on the same track may even be overwritten and destroyed. Caution must also be observed when using both GDWRT and RED256 from the same program, since the write check operation of the GDWRT routine reads data into the buffer used by the RED256 routine, thereby destroying data previously present in the buffer. GDWRT performs the error detection and correction steps outlined above the returns with the zero flag true whenever the write operation was successful. If some write error occurs, the zero flag will be false (a JNZ instruction will jump). The contents of the 8080 registers are destroyed by GDWRT in either case, and only the Z flag returns with valid information.

RED256

RED256 (for REaD a 256 byte block) reads a data block in from a specified disk, track and sector. When RED256 is called, it initiates the read operation, and assumes that the disk, track and sector of the data block to be read are stored in the memory locations DSKWNT, TRKWNT and SECWNT, respectively, with the same restrictions on the disk track and sector values as in the GDWRT routine. After being called, RED256 will return to the calling program after a delay of from 0.1 to 3 seconds. Upon returning, the zero flag is set (true) to indicate that the data block read in had the proper cyclic redundancy check bytes, and that the read operation was successful. The zero flag is cleared (false) to indicate that the read failed to produce the proper check value even after the multiple reads and head jogs described in the read error correction section. After a successful read, the block of data read will be present in the 280 byte buffer area from RDFLNM to RDFLNM+279, in the format shown in table 2. Because the buffer area is also used to check for errors during a write operation, as noted previously, the data read from the disk by a call to RED256 should be copied into another buffer area if the data is to be kept beyond the time of the next data block read or data block write. None of the 8080 registers are sacred during a call to RED256, and the programmer must take this into account.

The three entry points described, DKINT, GDWRT and RED256, are sufficient for all operations involved in any data base management or operating system which uses the data block format of table 2. However, the personal computing enthusiast (by definition never satisfied) may wish to use different data formats or different disk drives.

Listing 1, continued:

```

E142 3A47E2      LDA      TRAKNO
E145 77          MOV      M,A      ; STORE THE OLD CURRENT TRACK NUMBER
E146 D3F6      DSKN1:  OUT      HDUNLD  ; FORCE THE HEAD TO UNLOAD
E148 E1          POP      H
E149 F1          POP      PSW      ; RETRIEVE THE OLD DATA
E14A 77          MOV      M,A      ; STORE THE NEW NUMBER AS THE OLD ONE
E14B 4F          MOV      C,A      ; BC=NEXT DISK
E14C 0F          RRC
E14D 0F          RRC
E14E 0F          RRC      ; A=NEWTRACK*32
E14F 3246E2     STA      FDBUF    ; CONSIDER THIS THE NEXT DRIVE COMMAND
E152 D3F3      OUT      FDOUT    ; SET UP AHEAD OF TIME TO DEGLITCH
E154 DBF1      IN       FDSTAT   ; CHECK THE STATUS OF THIS DISK
E156 E604      ANI      FDRDY    ; IS IT LOADED AND USEFUL?
E158 C285E1     JNZ      DSKR1   ; NO, TELL USER ABOUT IT!
E15B 09          DAD      B
E15C 23          INX      H
E15D 7E          MOV      A,M      ; THIS IS WHERE THE DISK WAS LEFT
E15E 3247E2     STA      TRAKNO  ; DECLARE IT THE CURRENT COUNT
E161 FE4D      CPI      77      ; 0-76?
E163 F8          RM

E164 DBF1      GTRK0:  IN       FDSTAT
E166 E604      DSKN2:  ANI      FDRDY
E168 C284E1     JNZ      NODSK   ; NO DISK
E16B CDA3E1     CALL    TRAKIN  ; GET TRACK
E16E CDA3E1     CALL    TRAKIN  ; GET TRACK
E171 CDA3E1     CALL    TRAKIN  ; GET TRACK
E174 CDA3E1     CALL    TRAKIN  ; GET TRACK
E177 CDB2E1     TRZLP:  CALL    TRAKOT ; GET TRACK
E17A DBF1      IN       FDSTAT   ; CHECK STATUS
E17C E601      ANI      TRKZRO  ; IS TRACK ZERO?
E17E C277E1     JNZ      TRZLP  ; YES, ZERO TRACK
E181 3247E2     STA      TRAKNO  ; STORE TRACK
E184 C9          NODSK:  RET

E185 CD1000     DSKR1:  CALL    ERTYP  ; DISK NUMBER >7
E188 03          DB      ERDKNM
E189 C9          RET

E18A 3EFF      INIT:   MVI      A,255
E18C 323AE2     STA      LSTDSK
E18F 3C          INR      A
E190 F5          INI1:   PUSH     PSW
E191 CD9CE1     CALL    DKINT   ; GET TRACK
E194 F1          POP      PSW
E195 3C          INR      A
E196 FE00      CPI      DSKLIM
E198 C290E1     JNZ      INI1
E19B C9          RET

E19C CD2BE1     DKINT:  CALL    DSKNUM
E19F CD64E1     CALL    GTRK0
E1A2 C9          RET

; THESE ROUTINES MOVE THE TRACK:
; THESE ROUTINES MOVE THE HEAD IN (TOWARDS HUB) OR OUT (TOWARDS CIRCUMFER)
; ONE TRACK FOR EACH CALL,9          CALL LASTS 10 MILSECS
E1A3 2147E2     TRAKIN: LXI      H,TRAKNO
E1A6 34          INR      M      ; COUNT UP ONE TRACK
E1A7 3A46E2     LDA      FDBUF
E1AA F604      ORI      DIRIN
E1AC 3246E2     STA      FDBUF
E1AF C3BE1      JMP      TRAKMV
E1B2 2147E2     TRAKOT: LXI      H,TRAKNO
E1B5 35          DCR      M      ; COUNT DOWN THE TRACK
E1B6 3A46E2     LDA      FDBUF
E1B9 E6FB      ANI      DIROT   ; CLEAR FOR AN OUT MOVE
E1BB 3246E2     STA      FDBUF  ; SAVE IT
E1BE F610      TRAKMV: ORI      STEPP  ; SET THE STEP BIT
E1C0 D3F3      OUT      FDOUT  ; AND THE START BIT
E1C2 E6EF      ANI      STEPM  ; CLEAR STEP BIT
E1C4 D3F3      OUT      FDOUT
E1C6 CDCAE1     CALL    TENMIL  ; WAIT TEN MILSECS
E1C9 C9          RET

; THIS ROUTINE IS JUST A SOFTWARE 10MS WAIT
E1CA 01FF03     TENMIL: LXI      B,1777H
E1CD AF          XRA      A
E1CE 0B          TMLP:  DCX      B
E1CF B9          CMP      C
E1D0 C2CEE1     JNZ      TMLP
E1D3 B8          CMP      B
E1D4 C2CEE1     JNZ      TMLP
E1D7 C9          RET

; THIS ROUTINE GETS THE DESIRED TRACK (FOUND IN TRAKNT)
E1D8 3A48E2     TRKGET: LDA      TRKNT  ; GET THE DESIRED TRACK NUMBER
E1DB FE4D      CPI      77
E1DD F0          RP      ; ERROR
E1DE 2146E2     LXI      H,FDBUF
E1E1 7E          MOV      A,M
E1E2 E6FE      ANI      NOT WSEL
E1E4 77          MOV      M,A
E1E5 3A48E2     LDA      TRKNT
E1E8 DE2C      SBI      540
E1EA F2F1E1     JP      TRKG1
E1ED 7E          MOV      A,M
E1EE F601      ORI      WSEL
E1F0 77          MOV      M,A

E1F1 3A47E2     TRKG1:  LDA      TRAKNO  ; GET THE PRESENT TRACK NUMBER
E1F4 47          MOV      B,A      ; B IS WHERE WE ARE AT
E1F5 3A48E2     LDA      TRKNT    ; A IS WHERE WE WANT TO BE
E1F8 B8          CMP      B
E1F9 C8          RZ      ; TADA... WE ARE THERE
E1FA F203E2     JP      INGIN    ; A>B THEN GO IN
E1FD CDB2E1     CALL    TRAKOT  ; A.<B, THEN OUT
E200 C3F1E1     JMP      TGETLP
E203 CDA3E1     INGIN:  CALL    TRAKIN
E206 C3F1E1     JMP      TGETLP

; THIS ROUTINE LOADS THE HEAD AND WAITS 30 MS FOR SETTLING

```

Listing 1, continued:

```

E209 DBF1 HEDLOD: IN FDSTAT
E20B E620 ANI HEADLD
E20D D3F5 OUT LOADHD
E20F C8 PNZ
E210 C0CAE1 CALL TENMIL
E213 C0CAE1 CALL TENMIL
E216 C0CAE1 CALL TENMIL
E219 C9 RET
E21A DBF1 SGTGET IN FDSTAT
E21C E610 ANI INDX
E21E C21AE2 JNZ SGTGET ;WAIT FOR THE INDEX PULSE
E221 CA19E2 LDA SECWNT ;GET THE DESIRED SECTOR
E224 E61E ANI 30
E226 17 MOV B:A
E227 DBF1 SCTL2: IN FDSTAT
E229 E608 ANI SCTL2
E22B C227E2 JNZ SCTL2
E22E 05 DCR B
E22F F8 PH
E230 DBF1 SCTL2: IN FDSTAT
E232 E608 ANI SCTL2
E234 CA19E2 JZ SCTL2 ;WAIT FOR END OF SECTOR PULSE
E237 C227E2 JMP SCTL2

;ALL THE FOLLOWING MUST BE PM
; ALL DATA FILES, LINKED OR CONTIGUOUS, SHALL HAVE THE FOLLOWING FORMAT:
;
; BYTE 0-8 FILNAM:EXT. 9 BYTES ASCII FILE NAME ON EACH BLOCK
; BYTE 9 DEVICE NUMBER, IGNORE
; BYTE 10,11 CURRENT TRACK (3-76), SECTOR (0-30; EVEN) OF THIS BLOCK
; BYTE 12,13 TRACK SECTOR OF PREVIOUS BLOCK (0,0 IF THIS IS FIRST)
; BYTE 14,15 TRACK SECTOR OF FOLLOWING BLOCK (0,0 IF THIS IS LAST)
; BYTE 16-19 FUTURE USE
; BYTE 20 BYTE COUNT OF INCOMPLETE DATA BLOCKS
; BYTE 21-276 256 DATA BYTES
; BYTE 278-279 CYCLIC REDUNDANCY CHECK BYTES

E23A 00 LSTDISK: DB 0 ;NUMBER OF THE LAST DISK TO BE USED
E23B 01 DSKTRK: DS DSKLIM ;UP TO 8 DISKS WITH SEPERATE TRACKS TO KEEP
E23C 0000 SCTRMT: DH 0 ;TEMPORARY POINTER TO NEXT SECTOR
E23D 00 REDTPV: DB 0 ;COUNTER FOR 2 ATTEMPTS
E23E 00 FDBUF: DB 0 ;BUFFERE TO THE FD CONTROLS
E23F 00 TRFIND: DB 0 ;CURRENT TRACK
E240 00 TRFINT: DB 0 ;DESIRED TRACK
E241 00 SECWNT: DB 0 ;DESIRED SECTOR NUMBER
E242 00 DSIWNT: DB 0 ;DESIRED DISK NUMBER
E243 01 WPTBUF: DS 16 ;128 BITS OF 8 SYNC'S THE CLOCK
E244 81 WPSNCB: DB SYNCB ;SYNC BYTE GOES HERE
E245 46494C4E WPFILN: DB 'FILNAM :EXT' ;NAME OF FILE AS IT APPEARS IN THE DIRECTORY
E246 41404558
E247 54
E248 00 WPEV: DB 0 ;ONE OF 16 DEVICES OF 16 TYPES
E249 0000 TPYWPT: DB 0,0 ;TRACK, SECTOR THAT THIS WILL BE WRITTEN TO
E24A 0000 WPFILN: DB 0,0 ;TRACK, SECTOR OF THE LAST BLOCK BEFORE THIS
E24B 0000 WPFILN: DB 0,0 ;TRACK, SECTOR OF THE NEXT BLOCK AFTER THIS
E24C 00 PCNT: DS 4 ;RESERVE 4 BYTES FOR EXPANSION
E24D 00 WPCNT: DB 0 ;PERALLY ONLY USED BY THE BYTOT ROUTINES
E24E 01 WPCAT: DS 256
E24F 01 WPCHS: EQU WPCHS+2
E250 81 PDSNCB: DB SYNCB ;HERE FALLS THE INPUT SYNCBYTE
E251 46494C4E PPFILN: DB 'FILNAM :EXT' ;SAME AS DIRECTORY ENTRY
E252 41404558
E253 54
E254 00 PDEV: DB 0 ;INCONSEQUENTIAL
E255 0000 TPYREC: DB 0,0 ;THIS MUST AGREE WITH THE TRACK, SECTOR DESIRED
E256 0000 PPFILN: DB 0,0 ;LAST TRACK SECTOR READ
E257 0000 PPFILN: DB 0,0 ;NEXT TRACK SECTOR READ FOR THIS FILE
E258 00 PCNT: DS 4 ;EXPANSION
E259 00 PDDAT: DS 256
E25A 01 PDCHS: EQU PDSNCB+LN256C
E25B 0000 ENO

BYTYP 002F CHV42 E0FC CHVLO E0F9 CHVSU E0F6
CHKHI E07E CHV1R E06F CLFLO E021 CD 0000
CONPR E0ED CIPIN 0004 CIPDT FFFB CVINT E13C
LTPR E22B CSKLI 0008 CSHPL E146 DSKN2 E166
VSHNU E12B CSHPI E185 DSHMN E24A EPDKN 0003
EPMS 0028 ERPEO 0002 ERTYP 0010 ERWPT 0001
FDBUF E246 FDIHM 00F4 FDOTN 00F4 FDOT 00F3
FDPY 0004 FCPED 00F0 FDSTA 00F1 FDMR1 00F0
FUPES 0002 GCHWR1 E001 GDMRT 0000 GTPX0 E164
HVNJL 00F6 HEACL 0020 HEDLO E209 HEX1 004F
HEXCH 0043 INDX 0010 INGIN E203 IN11 E190
INIT E18A INPOR 00F0 LN2SN 0117 LNBUF 0100
LOADH 00F5 LSTD5 E23A NMTYP 0036 NODSK E184
JTPOP 00F0 PDCHE E43C PDCNT E38A PDDAT E38E
RDDEV E17F PDLN E376 PDSNC E375 PE025 E085
PECL0 E0CF PEDON E0B4 REDTR E245 RFWLN E384
PRVLI E382 SCTLG E21A SCTL2 E230 SCTLP E227
SCTNS E242 SCTL 0008 SECWN E249 SPTPR 00F3
SPTPS 00F2 SPTS0 00F2 SPTTF 00F1 STEPM FEF
STAPP 0010 SYNCB 0081 TENM1 E1CA TGETL E1F1
TNLP E1CE TRPK1 E1A1 TRPKM E18E TRPKN E247
TRPK0 E182 TRPK E0A8 TRPK1 E1F1 TRPK0 E1D8
TRKRE E380 TRPKM E248 TRKWP E266 TRKZR 0001
TRV3P E0A5 TRZLP E177 TXYTP 0055 UNSRF 0002
WSEL 0001 WFLN E26A WRCHP E372 WPCNT E270
WPDAT E271 WPEV E255 WPFILN E25C WRSNC E258
WPT25 E015 WPTBU E24B WPTGR 0008 WPTLO E058
WPFILN E268 ;:TRA 0016

```

The following subroutines, which are contained in this software, are useful for such purposes.

WRT256

This subroutine performs all of the write operations: generate preamble, sync byte, cyclic redundancy calculation, head load and seek, and data write. This is all that is done; no verify operation is performed, and the write is only performed once per call of WRT256.

REDONC

This subroutine performs all of the read operations: head load and seek, data read, cyclic redundancy check generation and comparison. Control is returned to the calling program with the Z flag set if the calculated cyclic redundancy check equals the value read from the disk. The read is performed only once per call of REDONC, regardless of whether the checking information indicates a proper read or not.

DSKNUM

In a multiple disk drive system, when changing from a read or write operation on one disk to a read or write operation on another drive, this routine must be called. The number of the next drive to transfer data to or from must be in the accumulator (a value of 0 to 7) when DSKNUM is called. If the number of the next drive to be used is different from that of the last drive, then DSKNUM will unload the head of the last drive, store the number of the track where the head of the last drive is positioned, enable the next drive, and recall the track where the head of the next drive is positioned. If the next and last drives are the same, no action is taken.

CHKSUM

This routine is entered with the HL register pair pointing to the first byte of some block of data whose cyclic redundancy check is to be calculated. The BC register pair contains some number from 1 to 65,536, indicating how many bytes are used in the calculation. The 16 bit check value is returned in the DE register pair. The program used to calculate this binary polynomial is adapted from the Intel Users Library (program 80-41).

TRKGET

The desired track, from 0 to 76, is stored in the byte labeled TRKWNT. When TRKGET is called, the selected disk (select-

ted by DSKNUM above) will move its data transfer head in or out until it is over the desired track. The desired track number is compared to 44, and the proper write current is selected on the WRITE CURRENT SELECT line of the interface. Upon returning from TRKGET, the head is at the proper track and the required head settling time has transpired. Calling TRKGET when the head is already at the proper track causes an immediate return; this will not significantly slow down the calling program.

TRAKIN

Calling this routine moves the data transfer head of the selected disk inward one track and increments the track counter. A 10 ms track movement delay occurs before control returns to the calling program.

TRAKOT

TRAKOT performs the inverse of TRAKIN by moving the head out one track, decrementing the track counter and delaying 10 ms before returning.

TENMIL

This is a software delay routine which delays 10 ms before returning. When using an 8080 or Z-80 system with a faster cycle time than 2 MHz, the delay loop counter should be changed. Because this software will not operate properly in memory with a cycle time longer than 500 ns, no values for slow memory need be given.

HEDLOD

This routine generates a head load pulse which causes the head of the selected disk drive to be loaded for at least 3 seconds following the call to HEDLOD. If the head was not loaded when HEDLOD was called, a delay of 30 ms occurs before control is returned to the calling program. If the head was loaded when HEDLOD was called, then the return is immediate.

SCTGET

SCTGET is called with the desired sector value in the memory location labeled SECWNT. The routine waits for an index pulse to occur on the selected disk drive, and then counts the sector pulses that follow. The first sector pulse following an index pulse is the start of sector 0, and the last pulse preceding an index pulse is the start of sector 31. Control is returned to the calling program within 15 to 25 μ s of when the leading edge of the desired sector pulse is found. Only registers A and B are used by this routine, allowing all other registers to be used by the calling program.

Conclusion

The software described in this article allows the advanced computer experimenter, who has implemented the economy floppy disk interface, to make good use of his/her floppy disk drive(s) for data storage and retrieval. These routines have been operational for well over one year, and for the last year, GDWRT, RED256 and DKINT have formed the heart of my 4 K byte (including data buffer areas) 8080 floppy disk operating system. This operating system allows machine language programs to be stored and retrieved, MITS BASIC programs to be stored and loaded (through the CSAVE and CLOAD routines), editing of text files (this article and my previous one were written and edited on the disk system) and batch stream processing (by assigning the keyboard input to be taken from a batch stream file on the disk). An unlimited number of files may be open for input or output simultaneously, requiring only a 280 byte buffer area for each file. BASIC programs can create, read, write and modify data files in ASCII or binary. All files can be specified with a 9 letter file name and a disk number.

The rest is up to you. This system can give you the extra bit of programming power you've been looking for! ■

NOTE: A floppy disk operating system is available for use with the interface described in February 1977 BYTE. The operating system runs in 4 K of 500 ns (or faster) user memory addressed at D000 hexadecimal.

The operating system is available on a diskette with two bootstrappable copies of FDOS on tracks 0 and 1, and the source code for the operating system is also supplied on the diskette.

Programs supplied on the diskette allow the user to reconfigure the FDOS to use his own IO devices and to store and run any of the users programs by name. The diskette and documentation are available for \$40 (Ohio residents add 5.5% tax) from K B Welles, 2623 Fenwick Rd, University Heights OH 44118.

Object Code in Machine Readable Form

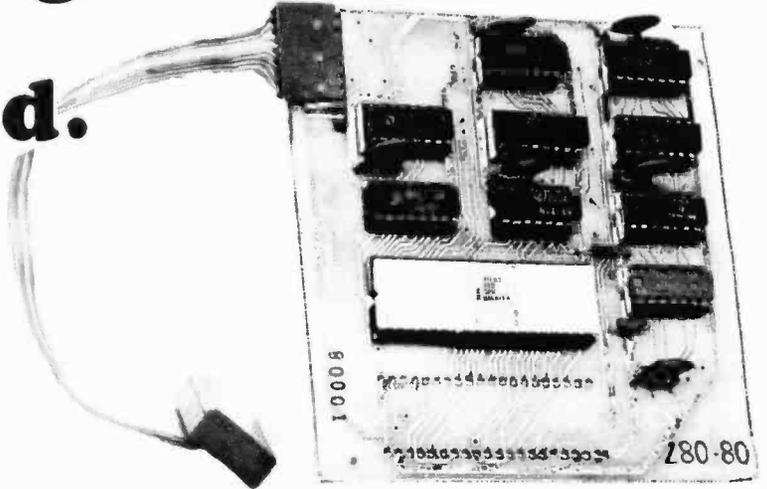
The bar code representation of Dr Welles' floppy disk software is expected to be prepared in time for the July issue of BYTE, where it will be printed as a small feature supplementing the information in this month's article.

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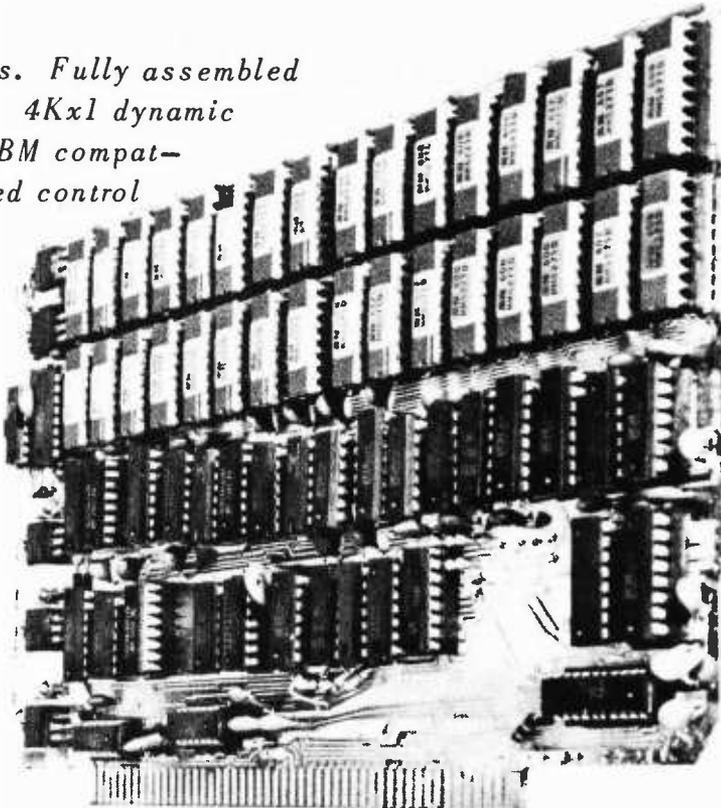
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Artificial Intelligence,

Part 2: Implementation

Michael Wimble
6026 Underwood Av
Cedar Rapids IA 52404

As described last month in part 1, there are five types of mutations that can be performed with the simulated evolution technique. A separate subroutine will be used for each mutation type. Four of the subroutines rely heavily on a subroutine which generates a random number between limits. (For those systems not already possessing random number generators, a box

accompanying this article gives an algorithm to produce pseudorandom numbers by the power residue method.)

Figures 1 through 9 are flowcharts of the basic modules which were extracted from a fairly sophisticated system of FORTRAN programs. These are intended to serve as a starting point for the reader in implementing his/her own program. If there is sufficient reader interest, I would be happy to program and publish program listings of implementations for one or more small system processors, in any popular computer language.

The rest of this article then is a description of the modules used to implement a 2 symbol gaming program using the artificial intelligence technique. If more detail is needed I suggest you look first into the book *Artificial Intelligence Through Simulated Evolution* by L J Fogel, A J Owens and M J Walsh (John Wiley and Sons, New York, 1966) or send questions and a self-addressed stamped envelope to me. Although there are many flowcharts and the technical jargon may seem complex at times, I wish to emphasize that the programming is simple and the technique can be implemented on any personal computing system with sufficient memory.

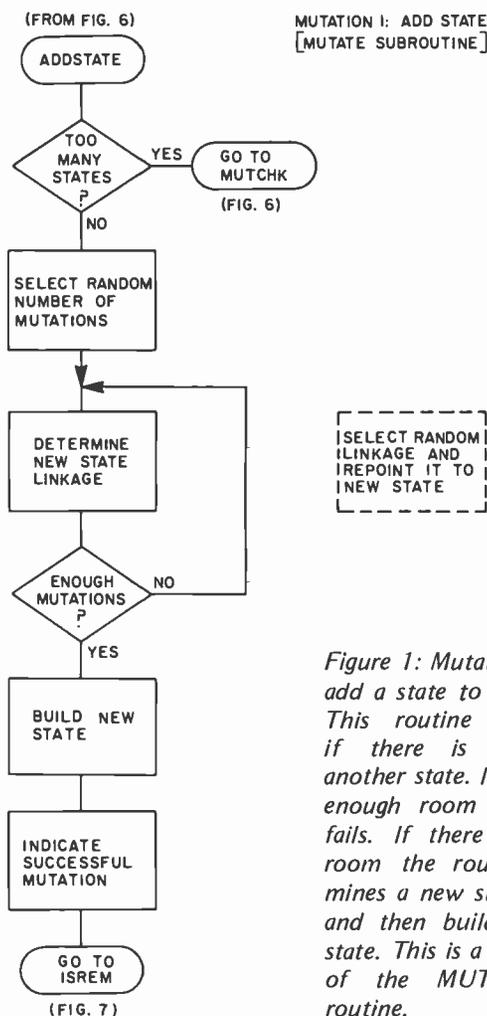


Figure 1: Mutation type 1, add a state to the model. This routine determines if there is room for another state. If there isn't enough room the model fails. If there is enough room the routine determines a new state linkage and then builds the new state. This is a component of the MUTATE subroutine.

Mutation 1 – Add a State

Figure 1 describes the first mutation type. As shown, the number of states is first compared to some maximum number. Most programs will have a fixed amount of memory allocated for containing the model, and so a check is made to see if any memory is yet available for expanding the model. The 2 symbol version previously discussed can conveniently be implemented using two bytes per state with the maximum number of states being 127. The programmer must determine the internal representation of the model and the amount of memory available and then set a variable to represent the maximum number of states the model can hold.

An Evolutionary Idea

The random number subroutine is called to provide an iteration counter. This counter is the number of randomly selected transitions that will be pointed at the new state. If no transitions were pointed to the new state then it would be an impossible state, ie: it could never be reached and could contribute no value to the model. Next, random transitions are changed to point to the new state which is to be created. Then, the new state is created. Finally ISREM is entered to remove any impossible states from the model.

Mutation 2 – Delete a State

Figure 2 describes the second mutation type. As with mutation type 1, a check is made of the number of states currently in the model. If there is only one state in the model, the mutation fails, since to delete the state would totally eradicate the model. If there is more than one state then the mutation will be able to proceed successfully.

Next, one of the states in the machine is randomly selected to be deleted. The only state that cannot be deleted with this subroutine is the state designated as the first state. To delete this state would result in the same mutation as mutation type 4 and is thus prohibited to this subroutine.

The actual deletion is accomplished in a rather roundabout manner. Every transition in the model is checked to see if it refers to the state being deleted. If so, it is repointed to some other random state number. The result of course is the creation of an impossible state, one that cannot be reached and is thus useless to the model. Entering the ISREM routine, as explained in the description of mutation type 1, will remove any impossible states from the model and so effect the actual deletion of the desired state.

Mutation 3 – Change a Transition

Figure 3 describes the third mutation type. The model is first checked for having more than one state. If there is only one state, then all transitions must necessarily point to that state, and so no change can be made and the mutation fails. Otherwise a

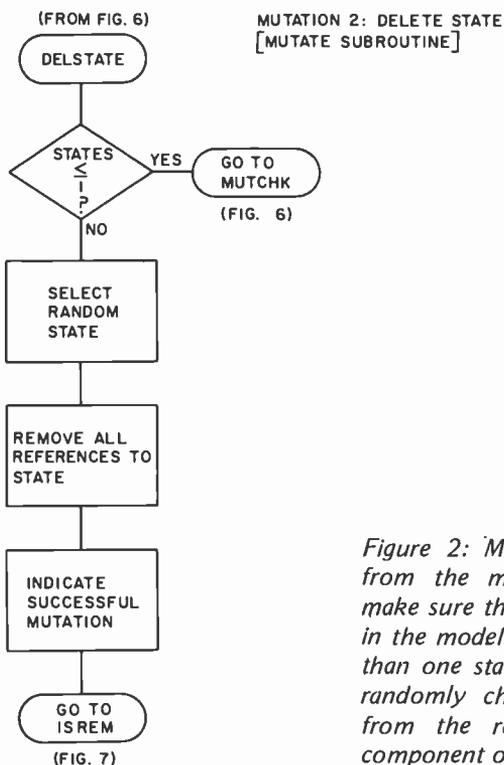


Figure 2: Mutation type 2, delete a state from the model. This routine checks to make sure that there is more than one state in the model or else it fails. If there is more than one state it deletes all references to a randomly chosen state thereby severing it from the rest of the model. This is a component of the MUTATE subroutine.

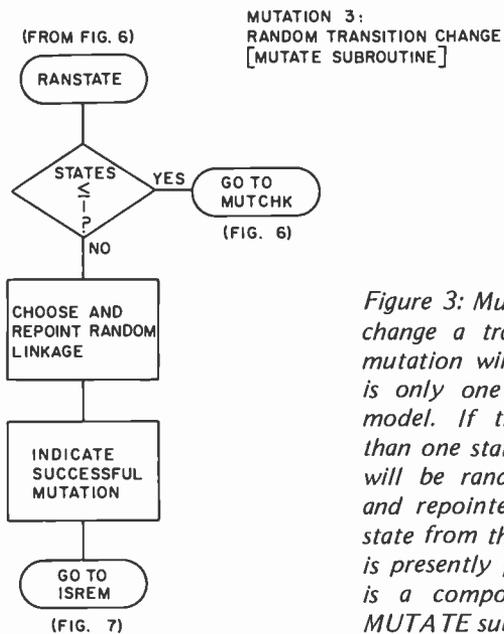


Figure 3: Mutation type 3, change a transition. This mutation will fail if there is only one state in the model. If there is more than one state a transition will be randomly chosen and repointed to another state from that to which it is presently pointing. This is a component of the MUTATE subroutine.

A Pseudorandom Number Algorithm

1. Test X, a 16 bit variable. If X is equal to zero, then set X to any number such as the time of day, or some other indeterminate number, and repeat step 1. This defines the initial seed of a pseudorandom sequence.
2. Multiply X by 259 to yield new value for X. Keep only the 16 least significant bits of the result. Increment X by 1.
3. If X is zero go to step 1.
4. Divide X by the input argument but do not destroy original value of X. The remainder of this result is a pseudorandom number between zero and one less than the input argument.

random element in the model is chosen and its transition pointer is changed to point to some other state. ISREM must be entered at the end of the routine since the changing of a transition may have resulted in the creation of an impossible state.

Mutation 4 – Change the First State Number

Figure 4 describes the fourth mutation type. Again, if there is only one state in the model, then it must necessarily also be the first state, and the mutation would then fail. Otherwise the first state number is set to any other state in the model. Also it is possible that an impossible state was created as the result of this mutation, so ISREM is entered to remove such impossible states.

Note that earlier discussion described the fourth mutation type as changing the current state number of the machine. The model must be driven by history from the first state when performing evolution, and it is seldom possible to work backwards from the current state number to determine the first state number. It is practical, therefore, to modify the mutation from change current state number to change first state number. The result of changing the first state number will usually end up changing the current state number.

Mutation 5 – Advance Model by One State

Figure 5 describes the simple but powerful mutation type 5. If the first state number is equal to the second state number of the model, then the mutation fails. Otherwise the first state number is set to the second state number.

Earlier discussion of mutation 5 described it as advancing the current state number by one. As with the case of mutation type 4, it is impractical to change the current state number directly, so the first state is advanced by one and evolution routines described later will run the model over its historical recall to result in the current state number actually being advanced by one state.

Mutation Control

Figure 6 shows the logic involved in controlling the mutation process. A random number is generated and used to select one of the five mutation types to be performed. Upon return from the mutation routine, a check is made for successful mutation. If the attempt was unsuccessful, a random selec-

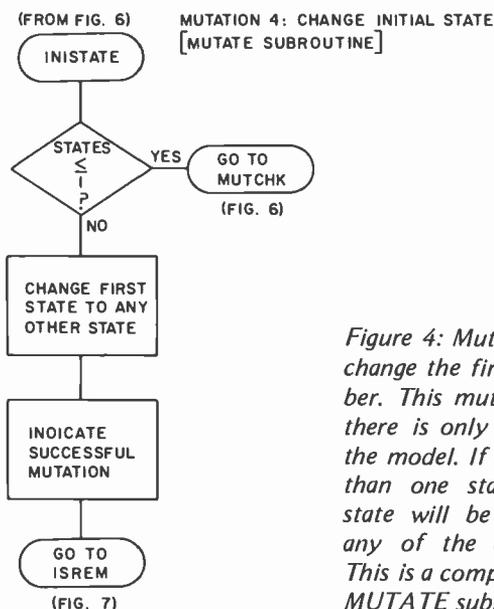


Figure 4: Mutation type 4, change the first state number. This mutation fails if there is only one state in the model. If there is more than one state the first state will be changed to any of the other states. This is a component of the MUTATE subroutine.

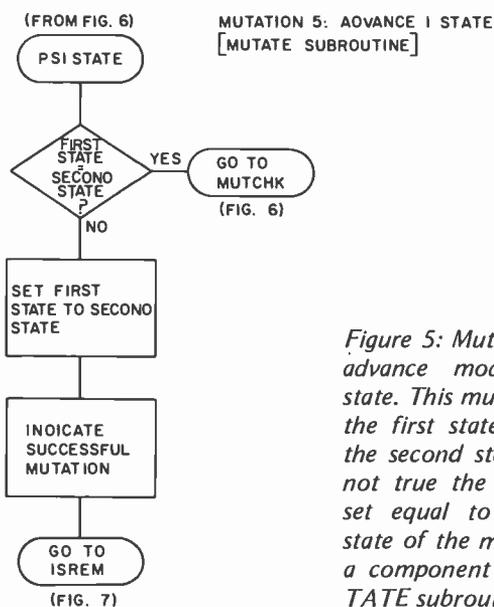


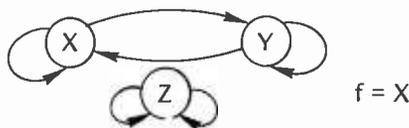
Figure 5: Mutation type 5, advance model by one state. This mutation fails if the first state is equal to the second state. If this is not true the first state is set equal to the second state of the model. This is a component of the MUTATE subroutine.

tion is again made until a successful mutation is finally made. Finally, the routine OPTIMIZE is used to perform deterministic optimization and to assign a value to the model.

For those who implement this artificial intelligence game and would like to improve the intelligence forming process, the mutation control routine of figure 6 offers the greatest potential for improvement. The mutation selection process shown is simple and effective, but is essentially a brute force method. One of the first improvements I made to the program was to allow dynamic altering of the probability of selecting a mutation type. That is, if ADDSTATE was called successfully and resulted in a model with a higher value than before, then the probability of selecting ADDSTATE in the future was increased slightly. Similarly, if any mutation type was performed successfully and decreased the value of the model, then the probability of selecting that mutation type again was made slightly less. The most sophisticated versions of this artificial intelligence program use the very artificial intelligence process to optimize the selection of mutation types, thus aiding the evolution process.

ISREM – Remove Impossible States

Figure 7 describes the subroutine that removes impossible states. An impossible state is any state that cannot be reached. For example, the model:



has the impossible state Z. An impossible state cannot contribute to the value of a model but can hinder the model. If a model already had the maximum number of states permitted, but five of these states were impossible states, then the ADDSTATE mutation could not be performed until some states were deleted to make room.

The flowchart of figure 7 will not remove all impossible states. For instance, if in the figure above, the current state was changed from X to Z, then X and Y would now be impossible states instead of Z. The subroutine described will only find those states which are impossible solely because they cannot be reached from another state of the current model. In practice, this has never yet failed to eventually find all impossible states.

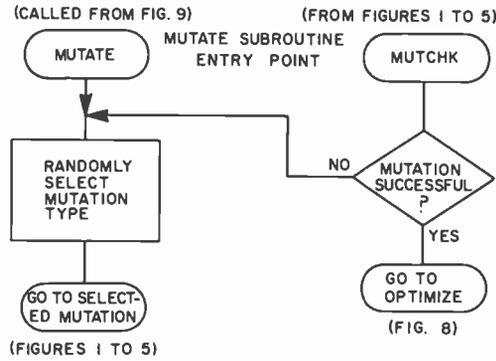


Figure 6: The MUTATE subroutine entry point. MUTATE chooses the type of mutation that will be performed using a pseudo-random number generator. At MUTCHK we check to see if the mutation was a success. If it was not then the MUTATE routine continues by choosing another mutation. (If at first you don't succeed . . .) If the mutation was a success then the optimization process cleans up the mutation by transforming it to its most usable form before returning control to the main program logic of PREDMUT.

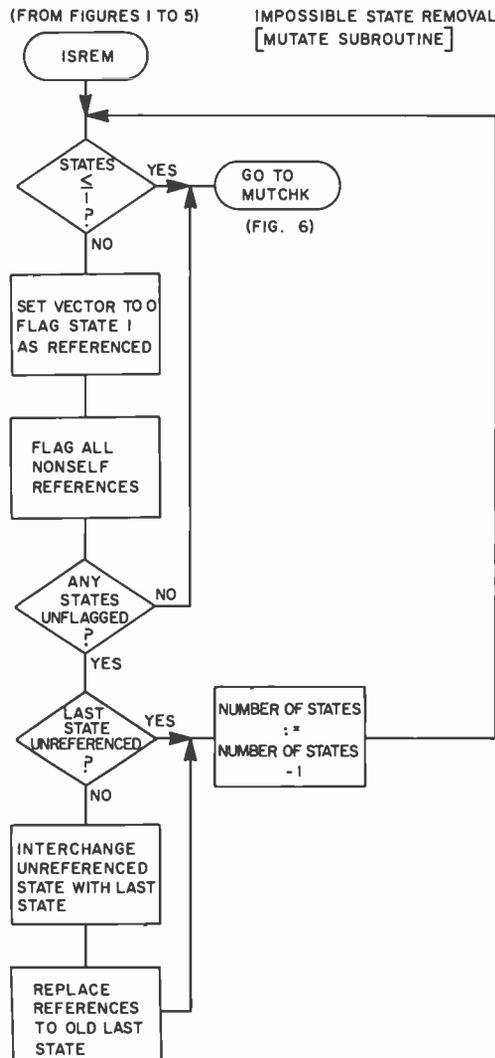


Figure 7: Routine ISREM removes any impossible states from the model. It performs this by checking all of the transitions to see that all of the states are referenced at least once. If it finds a state that is not referenced it will interchange that state with the last state of the model and delete the last position as valid. This is a house-keeping component of the MUTATE subroutine.

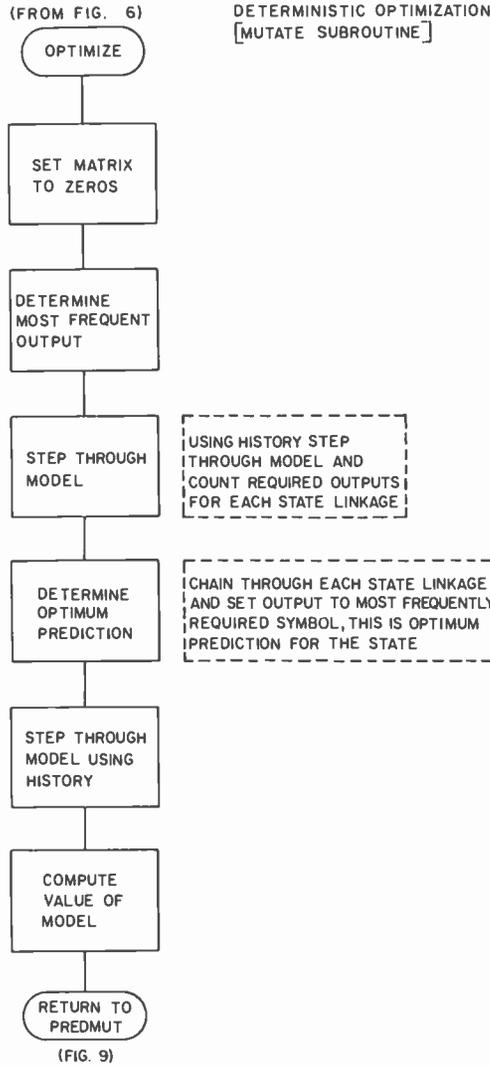


Figure 8: Routine OPTIMIZE determines the optimum prediction for each state of the new model. It will then evaluate the model using past history as a model to determine the value of the mutation. This is the final processing of the MUTATE subroutine of the program.

Table 1: The matrix used to evaluate the optimized form of the new mutation. Each predicted input symbol is matched against the actual input. The value for that combination is then added to the total value of the model. In this manner, using historical methods, the new model can be compared to the older model to see if it is more efficient.

prediction	actual occurrence			
	symbol ₁	symbol ₂	symbol ₃	...
symbol ₁	value ₁₁	value ₁₂	value ₁₃	...
symbol ₂	value ₂₁	value ₂₂	value ₂₃	...
symbol ₃	value ₃₁	value ₃₂	value ₃₃	...
⋮	⋮	⋮	⋮	⋮
⋮	⋮	⋮	⋮	⋮

Looking at the flowchart then, the process is simple. First a table is set to zero. Every transition is then examined. If the head of the arrow points to a different state than the tail, then the entry in the table corresponding to the state pointed to by the head of the arrow is flagged.

After looking at all transitions, the table is examined. If any state in the model remains unflagged then it is an impossible state. If the impossible state is the last state in the model, then one need only decrement the variable denoting the number of states in the model to result in the deletion of the state. Otherwise the impossible state is first swapped with the last state in the model before the variable is decremented.

OPTIMIZE – Evaluate New Model

Figure 8 describes the most important segment of the program. OPTIMIZE performs the deterministic optimization mentioned previously, and evaluates the model with respect to the goal. Optimization and evaluation provide the criteria for the evolution selection procedure.

OPTIMIZE is conceptually simple, although often large of implementation. The model is made to perform with historical data. As each input observation is fetched and the appropriate transition made, the required output is noted. Thus for each transition we have a table of the number of times each output should have occurred for perfect prediction. The output symbol that should have occurred most frequently for a transition is then made the output for that transition. If the transition was never used during this historically driven run, then the transition is made to output the most frequently occurring output symbol over all of history.

After optimization, the model is rerun using the same historical data. Now, however, the optimized model is evaluated in terms of its ability to achieve the goal provided. The goal is expressed in a matrix of the form shown in table 1.

For each transition the prediction and actual occurrence are used to extract a value from the table. The sum of these values is the value of the model.

For the purpose of predicting primes, the matrix is defined in table 2a, while for earthquakes the matrix is defined as in table 2b. You see then that each accurate prediction of an actual earthquake or no earthquake adds 1000 to the value for the model. To predict an earthquake when one does not occur adds 500 to the value of the model,

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(a)

prediction	actual	
	prime	not prime
prime	1	0
not prime	0	1

(b)

prediction	actual	
	earthquake	no earthquake
earthquake	1000	500
no earthquake	0	1000

Table 2: Two tables which illustrate examples of goals used to evaluate the models using historical data methods. Table 2a shows that when a correct guess of either prime or not prime is made a value of 1 is added to the total value of the state. Table 2b illustrates another manner of weighting answers. A correct answer receives a weight of 1000. Predicting an earthquake when there isn't one only adds a weight of 500 to the model. Not predicting an earthquake when there is going to be one adds nothing to the value of the model. This is a more subtle form of weighting since it not only allows a good and bad answer but also a not so good answer.

while missing an earthquake prediction adds nothing to the value of the model.

PREDMUT – Putting It All Together

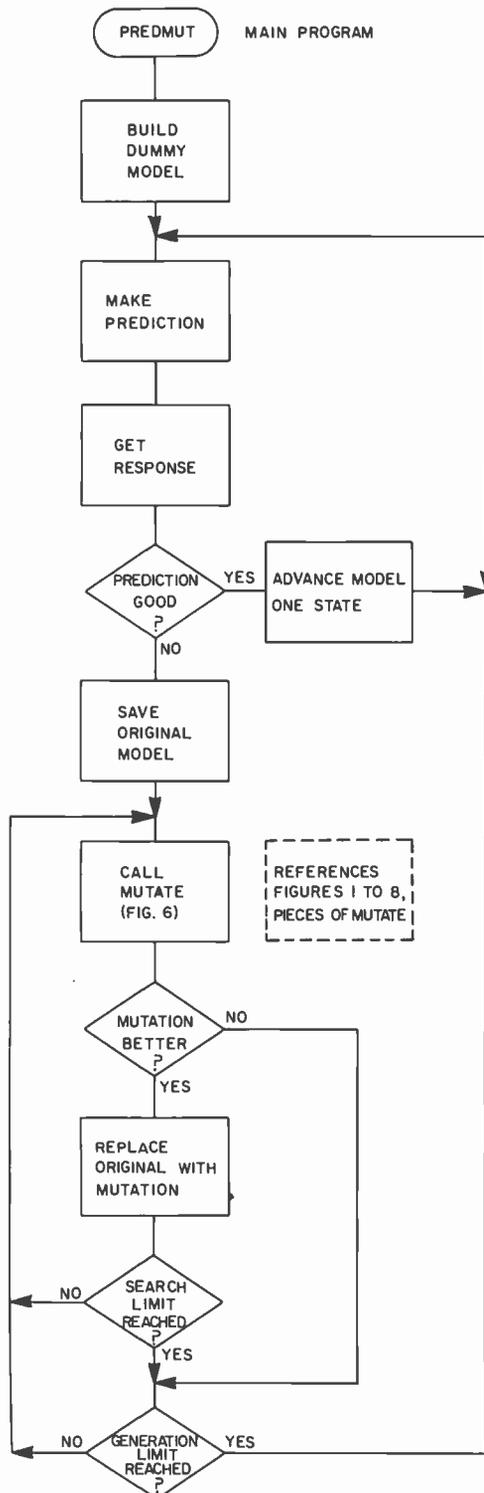
Figure 9 shows the mainline logic for the program. The flowchart should be obvious except perhaps for the use of the terms search limit and generation limit.

In order to prevent the evolution procedure from taking hours or days, it is constrained as to how long it can take. Each time an evolution cycle must occur, two variables are set. The search limit variable defines the number of times the parent may be serially mutated when searching for a better offspring. The generation limit variable defines the number of offspring to be generated.

The result is that a fixed number of offspring are generated. Further, each offspring can undergo a maximum series of mutations, since it is seldom that a single mutation results in a better model. To make the program more intelligent, one can increase these limit constants, but the result is a greater amount of computer time required between responses.

Again, for those adventurous programmers who want to make their programs even smarter, other methods can be employed to determine dynamically how many offspring and how many mutations are to be performed. There are other advanced evolutionary techniques that can also be employed. Interbreeding, majority logic and second order pattern recognition are just a few terms describing the advanced tech-

Figure 9: This is the main control logic of the simulated evolution technique. It first builds a dummy model, predicts an event, and receives the response. It then performs repeated versions and variations of the model looking for the optimum model to predict the correct answers.



niques available in computer science literature.

In Conclusion

If you have the patience, sit down with a pencil and paper and attempt to perform the process I've just described. Many mathematical books and papers contain random number tables you can use. As a result you should quickly see how this mathematical technique results in the creation of a model of some process. From there it is relatively simple to program.

Do not be afraid to try your own improvements, but do so intelligently as you get a feel for the underlying processes. The process described herein is the basis for many computer versions of this artificial intelligence technique and has worked with varying degrees of success for a wide variety of goals. I might conclude from personal experience: Be careful how and to whom you expose this technique. There are many people who fear computers as they fear anything they don't understand and their enthusiasm for your creation may not match yours. With a good enough program model, however, you should be able to predict who these people will be. [*Hmm . . . our artificial life has defense mechanisms.*] ■

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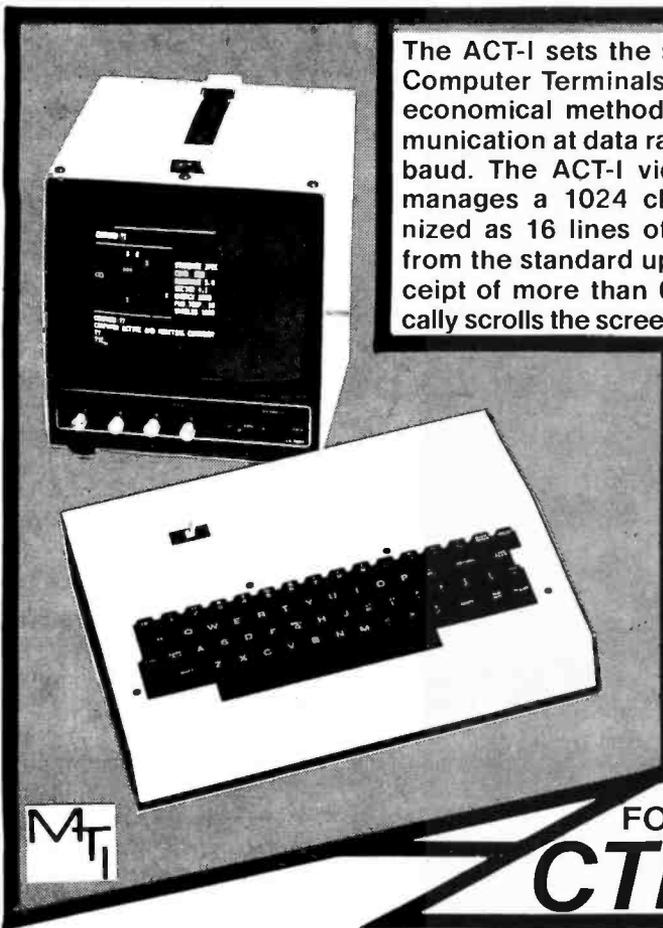
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Clubs and Newsletters

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Owners of KIM microprocessors should look into *KIM Users Notes*, published monthly by Eric C Rehnke, Apt 207, 7656 Broadview Rd, Parma OH 44134. The subscription rate is \$5 for the first six issues.

Alberta Microprocessor

David Lavers, a member of the recently formed Alberta Microcomputer Society, has been in touch. Right now information at this end is a bit sketchy, but you can find out more through their acting president Dwight K Soloman, c/o The Computer Hobby Shop, 4812 16th St SW, Calgary Alberta CANADA T2T 4J5, (403) 243-6776.

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South Florida Computer Group

The South Florida Computer Group has apparently grown to the status of a major computer club. Meetings are held in both Miami and Ft Lauderdale. Meeting times seem so flexible that it would be best to contact them directly to get their schedule. Their newsletter *I/O* has improved substantially over the past six months to the point that it is far more than just a review and

listing of club events. Volume 2, number 2 contains an in-depth look at resistors and capacitor color codes, and a piece on the new "minor loop" chips designed by Texas Instruments to be used with bubble memories.

To contact the South Florida Computer Group, write 1155 NW 14th St, POB 236188, Miami FL 33123, or phone (305) 324-5572.

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The Computer Hobbyist Group — North Texas

The Computer Hobbyist Group of North Texas is *the* place for computer buffs in the Dallas-Fort Worth area to develop and share an understanding of the hobby.

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Chicago Area Computer Hobbyist Exchange

Looking through *The Register*, newsletter of CACHE, I came across what may prove to be an interesting series on languages. In the coming months *The Register* is planning on a series of articles on BASIC, PL/I, SNOBOL, PASCAL, PILOT,

FORTRAN, LISP, TRAC, CASUAL and FORTH. This should be a step toward clearing up language questions and misconceptions. Write to CACHE at POB 36, Vernon Hills IL 60061. *The Register* is available for \$10 per year.

Washington Amateur Computer Society

The Washington Amateur Computer Society is a growing group of hobbyists in the nation's capitol. Meetings are held on the last Friday of the month in the second floor conference room of St John's Hall at the Catholic University of America. WACS is interested in exchanging newsletters with other computer groups. Correspondence should be addressed to Washington Amateur Computer Society, c/o 4201 Massachusetts Av, Washington DC 20016. ■

Conducted by
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Continued from page 74

blasted-IV bytes. They are great if you never want to do anything fancy, but... With the bus address coming from an extended micro-code, one may simply decode a la 8080 and use 8212s as your IO ports. Also note that for some instructions there are free bits, ideal for adding instructions.

Interrupts will be painful, as the system doesn't have a stack, unless you use one of the aforementioned free bits to implement one (hint, hint).

My reaction to your suggestion about emulation of a 360 is unprintable. *[Quite reasonable evaluation... CH]* If you must, however, don't do it with the 8X300. Use Motorola's 10800 series. It's faster (50 ns, YES, 50 ns), and in the end would be less work. See, it's optimized for 360 type systems. Leave the 8X300 for the job it was designed for, a microcontroller.

Which gets around to the point of this letter: As a controller, the 8X300 is unbeatable. You neglected to mention what I consider to be the best feature of the system (after its speed). Every instruction operates on a bit string. You specify the starting bit and the length in bits of the field in the byte that you want the instruction to operate on. One can also do n bit rotates on any register to register instruction. Thus, you can add two registers and rotate right n bits in 250 ns or move bits 5 to 2 from memory to IV byte bits 7 to 4 without affecting the rest of the bits. This applies to all instructions. Thus, the 4 port kit can easily act as a front end to a busy computer for 8 Teletypes. Or consider a floppy disk. At double density, the bit transfer rate is 500 kHz, or a bit every 2 μs, or a bit every 8 instructions. Actually, as shift registers are so cheap, one might as well use them and free the microcontroller for other tasks like file handling. I have SMS's floppy disk controller interfaced to my 8080. At \$640 it's not cheap, but they do provide the address and data bus on connectors for "maintenance." They are rather tight with the source code in the read only memories, though. It would be nice to have a parallel processing floating point processor when not doing disk transfers, particularly as it is practically free.

In short, the hobby world should take a long look at this device. But keep in mind that it was designed as a controller, not a computer. Its architecture is optimized as a bit banger, and it does that superbly. As an emulator, it's the pits: Use a 2901 and a 2909, or any of the other bit slices and save yourself a bunch of time, trouble and hassle. Do use the 8X300 as a peripheral controller,

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Finally, a note about Scientific Micro Systems. They are the only company to provide really complete documentation. Their products work as advertised. They

gave me the shirt off their backs in helping with the nuclear instrumentation. It is truly refreshing to deal with such a company. I urge you to drop them a note as the microcontroller is only one of a large group of components they make. All are so pleasant to use, you kind of forget about them, which is the highest compliment I know how to give to an IC. ■

A Critique of Self-Modifying Code

In February 1977 BYTE, page 132, we published an article by Charles Howerton, giving the software of a package of utility routines for an 8080 processor. Joseph Newcomer of Carnegie-Mellon University sends along this critique of the methods of coding the BARC routines. It should be noted that within the context of an interrupt free single process computer, the BARC routines will work as described. The subtleties occur in cases where interrupts are allowed or one desires to put the program into write protected or read only memory regions.

Joseph M Newcomer
Computer Science Dept
Carnegie-Mellon University
Pittsburgh PA 15213

Mr Howerton's article in the February BYTE would have performed a much greater service to the community of programmers if he had coded it to allow recursion and provided a few extra items of documentation. After having expended a great deal of effort and cleverness to fit his code into 256 bytes, he then compromises reliability by having it modify itself. The concept of code modifying itself is undoubtedly the worst single idea that neophyte programmers fall in love with. Since the most obvious thing in the world is to take his BARC package and put it in a PROM, my first critique is that he has

carefully made sure that is not possible! Not only is self-modifying code useless in the world of read only memory, but experienced programmers know that code which modifies itself is inherently harder to debug.

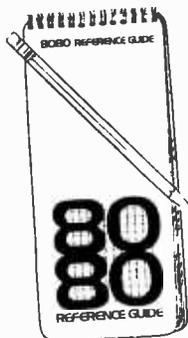
I should like to offer the obvious solution: Instead of storing the dispatch address in line in the code, store it out of line in the writable programmable memory. Thus, one must reserve a certain number of bytes for this address. However, since none of the routines in BARC call one another, only one dispatch address area is required. Then one only need do a "call indirect" through this address. Of course, the 8080 does not have a call indirect instruction, so one has to fake it; use a 3 byte dispatch area, and make the first byte the op code of a JMP

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instruction. This is, of course, "modifying code," but at least we now have the ability to put the program in read only memory.

```

writeable space:
GODO    JMP      Q'000000'
rom, typical call:
        SHLD    GOD0+1
        ...
        CALL   GOD0
    
```

Now, as to the issue of insufficient documentation: The functional documentation is quite clear, but the crucial piece of information that I find to be missing is that the routines in the BARC package should never be called from an interrupt service routine. This is because the routines are not reentrant (and my proposal of the dispatch vector in programmable memory does not change this); if they are active when an interrupt is taken, and the interrupt service calls them, the dispatch address will be changed. Consider the following sequence (events are listed):

```

CALL OCHR
...
CALL DASXR
...
CALL DASNT
--<interrupt>-----
...
CALL XCHR
...
<return from XCHR>
---<return from interrupt> ---
...
DASFC: CALL <something>
    
```

} Interrupt Changes
Dispatch Vector

where <something> is expected to be OCHR but in fact is XCHR. The difference in time between the setting of the dispatch address and the use of it is (if I counted correctly) 235 machine cycles: a very, very long time. Furthermore, this sort of bug is nearly impossible to locate at the "lights and switches" level because the act of single-stepping changes the relative timings and the error will not occur. Using a debugging system may have the same effect, and if it is a homebrew debugger which would use BARC, then it would destroy the very information it was trying to analyze. Absolutely nowhere in the article could I find any warnings about this! Only my experience was an indicator: Whenever static data, either code or otherwise, contains state information, look for cases in which routines which use it can be called recursively. Note that recursion does not have to be explicit; calling a routine from interrupt level where the routine was active at the time the interrupt was taken constitutes a recursive call.

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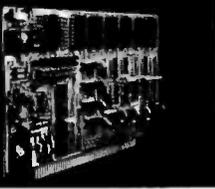
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The use of a single dispatch vector does make it possible, with care, to use BARC from an interrupt routine. Before invoking any BARC routines, push the current dispatch address onto the stack; after using BARC, pop it off. If BARC was active, the program has correctly preserved state; if not active, only a few cycles have been lost.

The requirement that the parameter lists be in line also seems to be a bad choice to me; it would be much better to simply store a pointer to the parameter lists, or even better pass a pointer in a register. Since speed was not of the essence and size was, it seems to me that something which reduces the size of the code at the call site would be more desirable. Furthermore, if the addresses or values of the parameters must change, in line parameters force the user to write self-modifying code. This means that the user cannot, after having developed some neat system, convert it to a PROM region of memory.

A good criterion for evaluating a program is: Can it be put in PROM? If, after separating out the data areas, it cannot be put in PROM, then from my experiences I consider it to be badly written, no matter how useful or amazing its functions may be. Then ask: Can it be called recursively? If not, do not call it from an interrupt routine if it can ever be called from outside an interrupt routine. And don't forget multilevel priority interrupts, which are very easy to handle on 8080 architecture. Do not call the routine from outside an interrupt context if it can be called from within one, which is the complementary condition. Note that these are not always obvious errors to detect, since the interrupt routine may call something which calls something which eventually calls something else; and one day the "something else" is modified so that it calls one of the nonrecursive routines. The scenario is now set for disaster: Let an interrupt come in at the wrong time and you are set for a long, tedious and probably unrewarding debugging session attempting to locate a source of "random" behavior.

Charles Howerton Replies

In response to the critique of BARC by Mr Newcomer of Carnegie-Mellon University, I should like to begin by saying that I agree with his comments without equivocation for the environment which he hypothesizes. I also agree that the concept of modifying running code is very "hairy." However, even Mr Newcomer's proposed "fix" requires the modification of running code.

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The BARC routines were written for newcomers to the computer programming field. It has been my experience that the greatest problems the beginner has in writing programs are manipulating the registers (hence, the register preservation facilities of BARC), and writing routines containing loops which are designed to perform the functions provided by BARC.

In addition, the newcomers to home computing or even industrial or business computing are rarely faced with programming an 8080 or Z-80 which is loaded with interrupt generating hardware. Those of us who work with interrupt driven hardware (and I do) would not use BARC for the simple reason that we are, presumably, sufficiently talented in programming arts that the creation of routines to perform equivalent functions is trivial.

As for embedding the parameters in line with the code, the one type of code that the average beginner writes well is straight line code with a few conditional jumps where decisions are required. Also, this is a fairly standard practice in large machine operating systems. Relative to the comment, "... if the addresses or values of the parameters must change ...", what is the difference whether they change in a work area or in the in line code? They must still be changed!

BARC was not designed to be called recursively; it was designed to be used as a programming tool by someone who was using straight line code to solve a problem. As for debugging it: it works.

Could BARC have been written to meet all of Mr Newcomer's criteria? Undoubtedly. However, not nearly as many functions could have been included in the same space.

Charles P Howerton
Digital Group Software Systems Inc
POB 1086
Arvada CO 80001 ■

Thanks to Charles Howerton
For BARC

Thanks for the Howerton article in your February 1977 issue, page 132, "Add Some BARC to Your 8080." Now I can finally get rid of the needless duplication and generally sloppy coding of utility routines in my programs and turn the job over to BARC. What a great time and memory saver!

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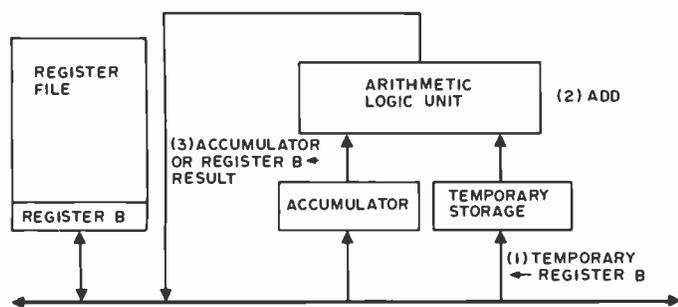


Figure 1: A block diagram with time notations for a sequence of events that might occur in a simple machine instruction such as: Add accumulator to register B. First (1) register B is put into a temporary storage area. Then (2) this storage area and the accumulator are added together by the arithmetic logic unit. The resulting answer (3) is stored in the accumulator or register B.

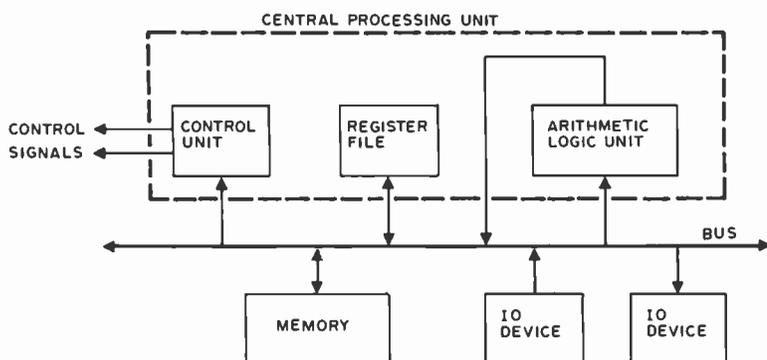


Figure 2: Block diagram of the architecture of a typical bus oriented digital computer.

What is Microprogramming?

When we consider the operation of a simple machine instruction, like add accumulator to register B on some computer, we often find that there is a sequence of even more elementary operations involved. For the example given, we may first have a transfer of data in register B to some temporary register in the arithmetic logic unit. Next, we may then perform an addition operation and finally, return the result of the operation to either the accumulator or register B. Figure 1 illustrates this sequence of operations.

Wilkes, an early pioneer in the field of computer design, called these elementary operations "microoperations." (See reference 1.) By this token, a single machine instruction, like the add described above, would consist of a microprogram of these microoperations. Microprogramming is, then, the implementing of control logic for a computer's instruction set through the ordered storage of processor control information.

Microprogrammable Computer Organization

Figure 2 is a simplified block diagram of the architecture of a digital computer. The organization of a microprogrammable computer differs from that of a nonmicroprogrammable computer in the design of the control unit. The nonmicroprogrammable

machine uses a hardwired control unit. All control lines are fixed and cannot be changed easily. On the other hand, a microprogrammable machine uses a changeable microprogram in implementing the control unit and thus by changing the microprogram, the machine can be altered within certain limits of its design. Let us now take a look at the control unit of a microprogrammable computer and figure out how it works.

A typical microcontrol unit would consist of a mapper, a microsequencer, a microcontrol storage and a decoder. The last item, the decoder unit, is optional and may not be found in some machines. The interconnections between these units are shown for a typical design in figure 3.

In operation, a machine instruction is fetched from main memory and is stored in the instruction register. The mapper converts this machine instruction into the starting address of the microprogram routine which is supposed to execute the instruction as a sequence of microoperations. This address is passed on to the microsequencer whose job is to step through the microprogram. As each microinstruction is read out, the decoder translates it into control signals for the various control lines.

Originally, the mapper was implemented using a decoder tree made up of discrete logic gates. Nowadays, array logic blocks in the form of read only memory and so called programmed logic arrays are used for this purpose. (It should be noted that programmed logic arrays are especially suited for this task. Read only memories contain many more bits than are needed, and are thus more expensive than programmed logic arrays. The array is powerful enough to implement most functions needed and its lower cost makes it a very attractive candidate for the mapper.)

The function of the microsequencer is to provide a value for the next address of an instruction in the control memory. It can be thought of as having a microprogram counter and additional logic to test for conditional branches. Thus, in its simplest form, it could just be a presettable counter with associated circuitry for performing branches and conditional tests. However, most commercially available micro-

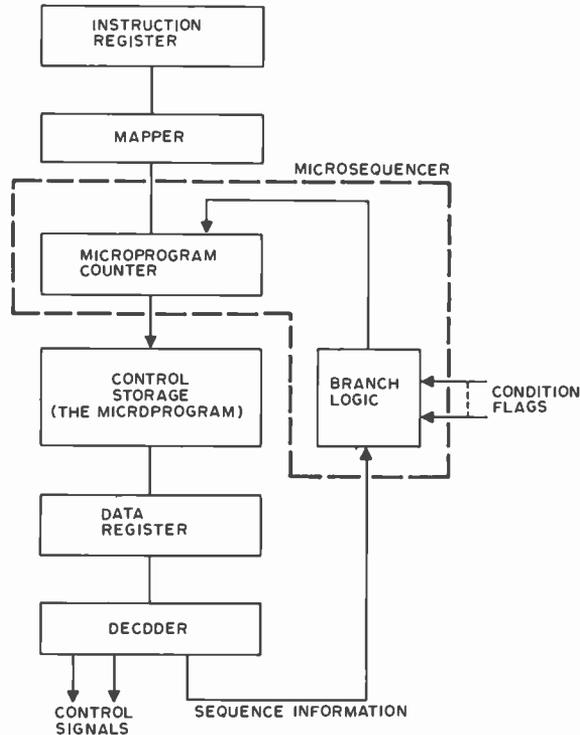


Figure 3: Block diagram of the connections between various parts of a microprogrammed central processor. The typical processor consists of a mapper, a microsequencer, control storage and an optional decoder.

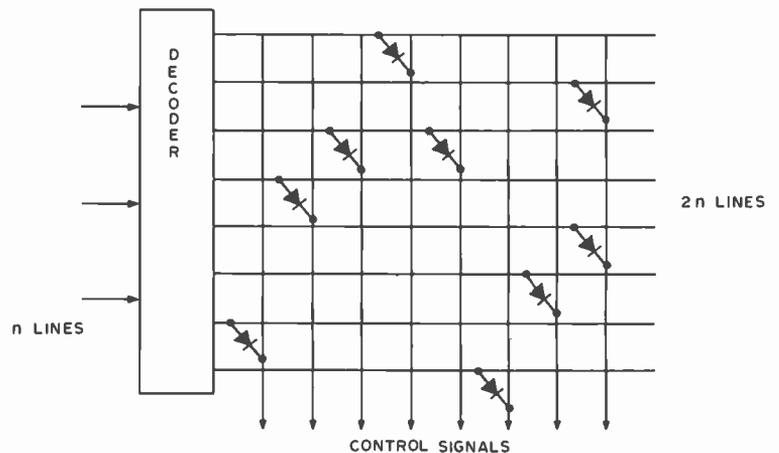
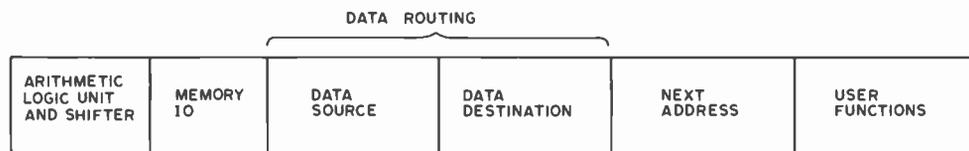


Figure 4: Typical diode matrix. A binary number is input on n number of lines. It is decoded into 2^n number of outputs. These lines are then decoded through the use of a diode matrix. The outputs of the matrix are at the control signals lines.

Figure 5: Typical microinstruction word format. The instruction allocations are generalized but are found in almost all such data formats. The word format itself can be of any length. The usual length is from 24 to 60 bits although a field width of 100 bits is used by IBM in some applications.



sequencers (such as the Intel 3001 or AMD 2909) are more sophisticated than this; some (eg: AMD 2909) even to the extent of having a built-in stack for processing microprogram subroutine linkages.

The microcontrol store is usually some kind of read only memory. Wilkes envisioned this as a diode matrix such as shown in figure 4. Of course, immense technological advances have been made since Wilkes' time and now the microcontrol store is usually implemented by read only memories, of which the discrete diode matrix can be thought of as the forerunner. In addition to read only memories, some microprogrammable computers have a form of programmable memory as part of their microcontrol store. This allows for dynamic changes of microprogramming which lead to an even more flexible and powerful machine. With such a configuration, the microprogrammer can easily rewrite, add or delete portions of the microprogram to suit the particular task at hand.

From a consideration of the microcontrol store, we next proceed to a discussion of the microinstruction. It is the microinstruction that forms the control mechanism which causes each data register change. A typical microinstruction word format is shown in figure 5. Generally, there has to be a field to control the arithmetic logic unit and the shifter, one for memory, and one for IO control. In addition, another field has to be reserved for information regarding the routing of data. Some kind of sequencing field which specifies the next microprogram address is usually also included. Finally, to

suit the architecture of a particular machine, there is a field left for user definable functions. These vary from machine to machine but would usually include conditional tests and branches. From this brief discussion, it should be apparent that there is no fixed width for the microinstruction. Indeed, it varies from 16 bits for the Signetics 8X300, chip, to 24 bits for the HP21MX minicomputer to 100 bits for various models of the IBM360. However, the width of the microprogram word in most small and medium size computers ranges from 24 bits to 60 bits.

If the microinstruction is wide enough, we can allocate a single bit to a single control line. In such a case, the microinstruction is said to be unpacked or horizontal. However, if we want to save control memory space, we may want to encode the data so as to compress the word width. An external decoder can then be used to recover the data. This is the packed or vertical format. A machine seldom takes on a fully packed or unpacked format for its microinstruction. Instead, most machines have microinstructions which lie somewhere between the two extremes; some fields are encoded while others are not.

When deciding on the width of the microinstruction, several factors have to be considered. The first and most obvious is cost: The wider the microinstruction, the higher its cost. This is so because memory is more expensive than a decoder. The penalty paid for having a vertical or packed format is a decrease in speed and flexibility. In the case of a horizontal machine, there are separate bits controlling the individual lines. Thus, there can be more parallelism in the control as more resources can be controlled simultaneously. However, for this same reason, horizontal machines are much harder to program. The microinstruction set of vertical machines resembles the assembly languages of minicomputers.

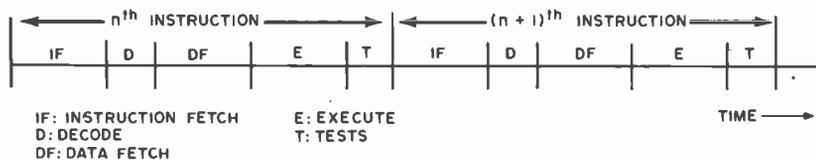


Figure 6: Timing diagram for the sequence of interpreting an instruction. This process is divided into five stages: instruction fetch, decode, data fetch, execution and testing. As soon as the testing is finished, the instruction fetch cycle is again encountered to start the next sequence.

Instruction Interpretation

Let us consider the sequence for the interpretation of an instruction. In figure 6, we see that the process of interpreting an

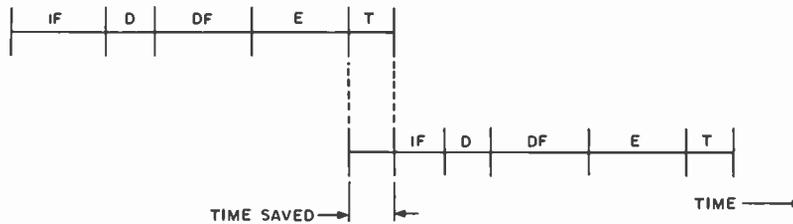


Figure 7: Timing diagram illustrating the time that is saved when implementing the changes of listing 2. This time saving occurs only when an interrupt or direct memory access request does not have to be processed. If a request does have to be processed, the timing diagram of figure 7 applies to the process.

instruction can be roughly divided into five stages. The first stage is that of an instruction fetch. The contents of the program counter is sent to the memory address register and a read memory command is initiated. The program counter is now incremented and the microcontrol unit waits until memory is ready with an instruction.

Once memory is ready, the instruction is loaded into the instruction register. This completes the first stage. On the time chart (figure 6), this corresponds to the segment IF. The mapper decodes this instruction into a microaddress which is passed on to the microsequencer. This is the decode phase, segment D on the time chart. As the microsequencer steps through the control memory, we have control signals coming out of the microcontrol memory. Depending on the instruction, we may need to initiate another memory read to fetch data. This would be segment DF. Once this is completed, we can proceed to instruction execution, segment E. Finally, upon completion of execution, a series of tests can be performed. These could include software tests for conditional branch hardware or software tests for interrupt, and hardware tests for direct memory access requests. The sequence for an add accumulator to register B (store results in accumulator) and skip if overflow may look like listing 1.

Of course the listing has to be coded into micromachine language form. DMA.SERVICE, INT.SERVICE and INSTRUCTION.FETCH would then be microprogram subroutines to service the various requests. A register, which holds data coming from memory, is assumed to be present. If no buffer register is used, then step 10 in the routine should be changed to:

(10) IR:= MEMORY DATA.

This will enable the mapper and load the microsequencer.

Why Microprogramming?

Before microprogramming was developed and firmly incorporated into computer de-

sign, most computer designers had to rely on multiphase "hardwired" logic for their design. Multiphase logic uses multiphase clocks to control the various register to register transfers and other functions. Hence, the designs are extremely complicated. Once the machine has been hardwired, it becomes virtually impossible to change the instruction set without redoing the design once again, ie: rewiring it.

Microprogramming overcomes these disadvantages and provides a means for obtaining relatively simple and flexible designs. To illustrate, reconsider the microprogram for the add instruction. If we look at the time chart of figure 6, we see that waiting for

Step	Instruction	Commentary
1	MAR:=PC	memory data counter:= program counter; [read memory]
2	PC:=PC+1	increment program counter;
3	[wait for memory];	
4	IR:=MDR	enable mapper, load microsequencer;
5	TEMP:=REG B	disable mapper;
6	ADD, ACC:=RESULTS	add and store results in accumulator;
7	IF OV=1 THEN PC:=PC+1	if overflow increment program counter;
8	IF DMA REQ=1 THEN JMP DMA.SERVICE	if direct memory access requested, go to routine;
9	IF INT REQ=1 THEN JMP INT.SERVICE	if interrupt requested, go to routine;
10	JMP INSTRUCTION.FETCH	fetch next instruction;

Listing 1: A program listing for the sequence: Add accumulator to register B, store results in accumulator and skip if overflow exists. Lines 1 through 4 are the instruction fetch routine. Lines 5 and 6 add register B to the accumulator and stores the result in register B. Line 7 checks for an overflow, and lines 8 through 10 check for interrupt and direct memory access requests. When the program listing is encoded into micromachine language form the DMA.SERVICE, INT.SERVICE, and INSTRUCTION.FETCH will become microprogram subroutines to service the various requests.

Step	Instruction	Commentary
9	MAR:=PC	read memory;
10	IF INT REQ=1 THEN JMP INT.SERVICE	if interrupt requested, go to routine;
11	JMP INSTRUCTION.FETCH+1	skip first step of instruction fetch;

Listing 2: Revisions of the program of listing 1 which allow a savings of time in the execution of the program. This savings is shown graphically in the timing diagram of figure 7.

memory to be ready with an instruction takes up a good portion of our time. Thus, to speed things, we may want to initiate a read memory instruction before we proceed to the various tests. Accordingly, we rewrite our microprogram with the modifications starting on line 9 shown in listing 2.

The new time chart obtained with this modification is shown in figure 7. Note that if no direct memory address requests or interrupts have been requested, then there is a savings in time. This is because we do not have to wait as long for the instruction fetch: a result of the early initiation of the read memory. If there has been an interrupt or direct memory access request, then no gain in speed would be obtained. However, since these requests are relatively rare, this new modification would result in an overall increase in speed.

If hardwired logic was used for the control unit, it would be very difficult to make the modification just described. Thus, we see that microprogramming is a very powerful tool in the design of digital computers. ■

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Continued from page 13

ELIZA INTEREST

John Aurelius' letter (March 1977 BYTE, page 16) struck a resonant chord. When I succeed in getting my own system, I have plans to implement a version of the ELIZA program. ELIZA was first described by Weizenbaum in the *Communications of the ACM (CACM)*, Volume 9, Number 1, January 1966, and has subsequently been discussed by Weizenbaum and others too numerous to mention. ELIZA is still alive and productive, although not necessarily as a psychotherapist. I am referring to a version of ELIZA implemented by Shapiro and Kwasny (*CACM*, 1975, 18, pages 459 to 462) as an interactive consultant for a timesharing system. ELIZA is a very general program whose specific personality derives from a script she is given. Shapiro and Kwasny reasoned that ELIZA would be a suitable mechanism for providing assistance to timesharing users. It was this application of ELIZA which intrigued me sufficiently to begin experimenting with her. Using their program as a point of departure, I have implemented my own (somewhat improved, I feel) version of ELIZA also designed to act as a timesharing system consultant. This program is now free of bugs (I think); however, the script is not yet very sophisticated, so a dialogue with the program is not yet rousing. I would be willing to provide a reasonably well documented listing to anyone seriously interested. Before the idly curious drop me a card requesting a copy, I should warn them that the program alone consists of 500 lines of SNOBOL 4 statements and requires 60 K words (the equivalent of approximately 360 K bytes) of storage to run on the Honeywell 635 computer for which it was written.

Now that everyone's enthusiasm has been dampened, I should admit that my choice of SNOBOL was expediency. ELIZA was originally written in LISP (also a memory hungry, slow, interpretive language). ELIZA requires the ability to search input text for certain desired content (pattern matching) and ELIZA's internal data structures are best represented as some form of list structure. These two attributes suggest that the most natural (but not necessarily most efficient) choice of a programming language is one that facilitates one or both of these ends. For implementing ELIZA on a small machine, I envision a relatively small core resident program to maintain the keylists and do the pattern matching, while the lengthy script is maintained as files on a secondary storage medium. These files would preferably be kept as random access files on a relatively fast device such as a floppy. A sequential device like cassettes might do, but I fear it would be intolerably slow. Unfortunately, ELIZA needs to keep a great deal of text at her fingertips (core resident in my SNOBOL program)

for matching against input and for reconstructing into output. Again, I would gladly exchange ideas with anyone interested in implementing such a program.

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University of Wisconsin
1025 W Johnson St
Madison WI 53706

COMMENTS ON TURING MACHINES

Jonathan K Millen's December 1976 article (page 114) on an actual hardware implementation of a universal Turing machine was very interesting. The relationship between the Turing concept of a universal machine (computer) and the capabilities of "real" computers continues to be ignored or misunderstood by many persons who think they know what a computer is. Millen's hardware project may help to enlighten these persons who view the Turing machine world as totally separate from the "real computer" world.

It may be of interest to note that Konrad Zuse appears to have investigated the Turing machine concept soon after World War II, hoping to find some ideas which could be used to simplify the construction of real computers. (My knowledge of this activity comes from a brief private conversation with Zuse at Los Alamos in June 1976.) He did not find any, of course, since we know now from experience that extremely simple machines in their primitive state are difficult to use in practice. Zuse's early work on computers was done without any knowledge of Turing's work, while it is known that von Neumann's input to early computers was not independent of such knowledge. (von Neumann became acquainted with Turing while Turing was at Princeton in 1937 and 1938 after the Turing machine work which was published in England.)

Millen makes some statements regarding the Busy Beaver Game that should be corrected. This game was invented by Tibor Rado and is described in the article "On Non-computable Functions" in the *Bell System Technical Journal*, Volume 41, May 1962. Millen has taken the liberty of adding an additional state to his Turing machines which he counts as a state for the Busy Beaver Game. This adds behavior that effectively eliminates his machine example from being considered in the game. It goes into a nonstop loop instead of halting. This also distorts the rules of the game. His 6 counting "4 state" machine is actually a 3 state machine by Rado's rules.

Millen then gives what he says are known results for 3 thru 7 states when he really should be saying 2 thru 6 states. In any case, the "known" results he mentions are still incorrect. The best results are as follows:

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Current Results (1975) Busy Beaver Game

(Millen's States:)	Correct Number of States	Busy Beaver Score:	Determined by:
(3)	2	= 4	T Rado
(4)	3	= 6	S Lin
(5)	4	= 13	A Brady
(6)	5	> 112	D Lynn
(7)	6	> 117	A Brady from 5 state result of D Lynn
	7	> 22,961	M Green
	8	> 3·(7.392-1)/12	M Green

While the summary I have shown here is in the process of being published, most of these results have been available in the open literature.

Allen H Brady
Univ of Nevada
Computing Center
POB 9068
Reno NV 89507

AUTHOR JONATHAN MILLEN REPLIES:

I am grateful to Dr Brady for his update on the Busy Beaver Game results. The terminal "copy" loop on state 4 of the example program was necessary because my universal Turing machine (UTM) has no automatic halt. A copy loop is an adequate substitute, because it is recognizable as such, and it does not change the contents of the tape.

Readers may be interested in the alternative Turing machine hardware realizations described in the following two references:

1. I Gilbert and J Cohen, "A Simple Hardware Model of a Turing Machine: Its Educational Use," *Proceedings of the ACM Annual Conference*, 1972.
2. Wakerly, J F, *Logic Design Projects Using Standard Integrated Circuits*, John Wiley and Sons, NY, 1976.

Jonathan Millen
66 Main St
Concord MA 01742

FURTHER SYS 8 EXTENSIONS

Readers who have been following the two recent articles in the January and February 1977 issues on improving SYS 8 Software Package 1 by Willard Nico may be interested to know about one of our products. This product is called Software Package 0.5.

This is a program in source code, plus a complete manual, for improving SYS 8 Package 1. In addition to the added commands and auto line capabilities discussed by Mr Nico, our program offers: insert, delete, and change string operations on a current line; string operations on a find first occurrence basis; page listing modes; reordering of line numbers; automatic tabbing; optional suppression of line numbers; and more. For the assembler, we add the following: octal numbers are accepted; a

global symbol table for often used symbols; an ASC pseudo op for including real time output lines; output of the symbol tables; an expandable table of pseudo ops; and still more.

The program also adds the capability to assemble programs in sections as they are read in from a mass storage device. This means that program size is no longer limited by the amount of programmable memory available for source code files!

The source code translates to 1.75 K object code bytes. The user edits this in with his existing SYS 8 Software Package 1, conforming it to the limits of his system. For this gem of a program, we charge a mere pittance, a token \$14.95.

Larry Weinstein
Objective Design Inc.
POB 20456
Tallahassee FL 32304

WINDMILL JOUSTING DEPARTMENT

I'm just getting started in this field and have enjoyed and benefited from the past six or seven issues of your publication. I will soon purchase and (I hope) have running a micro, both for personal use and as a fairly large data base device for a three man law office.

What prompts me to write this is the use of such expressions as "his/her" and "he or she" which appear with increasing regularity in your magazine. The zenith was reached in the February issue.

Let's be fair; you are leaving out a lot of readers. For example, there are certainly corporations and schools which own micros. Surely, the inclusion of only the masculine and feminine gender must seriously offend these neuter users. *[Let's start an "It" liberation front?]* Please, then, include "it" in such expressions as "Whenever the user finally gets his/her/its machine running . . ."

Secondly, there no doubt are some partnerships or associations that own or use micros. Can you imagine the chagrin of the members of these bodies not to be included in all your pronouns! This will necessitate your saying: "Whenever the user finally gets his/her/its/their machine running . . ."

Don't offend, for God's sake. And to heck with grammar or readability.

I am sure, if you really work at it, you could find even more ways to insure that your articles are hard to read — like,

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Space War (SPW) \$25
Deluxe Space War (DSPW) \$35
(Add suffix -T for Tarbell tape, or -P for INTEL hex paper tape.)

A complete source listing is available for an additional \$10 for either game.

Write for full description, or better yet, play a few rounds at your local computer store. But be prepared to stay a while. There is likely to be a line and you may become addicted.

MC and BAC accepted.



"owner/user," for example — we have to be accurate, right? Or "hobby/personal/recreation/small business" every time to describe (modify) the word "microcomputer." That'll really screw up the readers and give more money to the authors, too.

Be honest, folks; are you on some kind of a crusade? If not, please drop the abominable usage of English (even lawyers recognize the use of "his/her" and the like is bad) and get on with publishing valuable, concise and readable pieces for the benefit of your poor readers. If the gals are offended by the use of masculine pronouns, then fine — print 'em all in the feminine gender. Most men couldn't care less. But stop the foolishness, okay?

W C Welborn Jr
Caine and Welborn Law Office
2221 W Franklin St
Evansville IN 47712

Its, you? Gesundheit!

SOME COMMENTS ON MIKBUG

The following letter reaches readers in two parts. The main body of the letter is found here; the information in one paragraph of the letter is noted in "BYTE's Bugs" on page 160, and is not repeated here.

I was pleased to see John Rathkey's article "A MIKBUG Roadmap . . ." in February 1977 BYTE, page 96. The IO routines in this ROM are very useful and have saved me a lot of programming time.

The following comments and additions apply to table 1:

1. BADDR alters the contents of both A and B.
2. OUTHL and OUTHR both destroy the contents of A. Thus A must be reloaded if one desires to output both nybbles.
3. INHEX puts a hexadecimal digit in the right nybble of A.
4. OUT4HS outputs the four hexadecimal digits pointed to by X and X+1, then prints a space. X is incremented by 2 and the contents of A are lost.
5. OUT2HS also uses X as a pointer. X is incremented and A is altered.
6. OUTS is omitted from the list. This routine prints a space and begins at EOCC.

The routine labelled PSTR in listing 1 duplicates PDATA1 in MIKBUG. The only significant difference between the two routines is that PDATA1 uses the code 04 (EOT) to mark the end of the string rather than 00 as used by Rathkey.

The indexed mode JSR command can occasionally be used to save a few bytes of code when using the MIKBUG IO routines to and from the A register. Simply load the lowest address of the routines you will be using into X; the

desired routines can then all be addressed with the 2 byte indexed mode JSR rather than the 3 byte extended mode. Of course, this technique won't help if you want to use the routines that require X as a pointer.

The MIKBUG program uses a block of programmable memory from A000 to A049. Since this is usually provided by a 128 by 8 RAM chip, the remaining 53 bytes from A04A to A07F are available to the user for data or short programs. In addition, space for a stack is provided from A014 to A042. Since many programs won't require such a large stack, some of these bytes can generally be used for other purposes. The stack pointer is initially set to A042 and increments downward as more stack bytes are required.

You have a fine magazine; keep up the good work.

D B Brumm
dB Engineering
224 Hecla St
Lake Linden MI 49945

NOTES ON ARTICLE CONTENTS

I have been reading BYTE for about a year now. I would like to congratulate you on your very interesting journal. However, it disturbs me that you continue to publish articles on programming and construction techniques only for the more well known microcomputer chips such as the 8080 and 6800. I contend that articles on microcomputers based on the 1802 and the 6502 would also be beneficial as they are also in popular use.

Leonard P Jacobs Jr
USF #1570
Tampa FL 33620

We've had numerous articles on the 6502, already, starting with a review of the processor in November BYTE 1975 by Dan Fylstra, and continuing with several about the KIM-1 and its application or modification. In the near future we'll have David Brader's Komputar, a homebrew 6502 system plan. As for the 1802, or any other processor, what we print to a large extent depends on what people are doing, since the majority of BYTE articles are unsolicited contributions from readers actively engaged in experimentation. Based on recent data from readers, authors and manufacturers, there should be a bit of an upswing in 1802 awareness over the next few months.

THE EVOLVING LEXICON

A thorough answer to W Buchholz's question [February 1977 BYTE, page 144] about words that have passed from computer jargon into the general vocabulary would probably require a master's thesis.

The main reason is that several wholly new dictionaries have come on the US market in the past ten years or

so. Among them are: *American Heritage Dictionary* (three editions), *Doubleday Dictionary*, *Random House Dictionary* (two editions). Several older dictionaries have been extensively revised; set in new type, or both: *Webster's New World Second College Edition*, *Merriam-Webster Seventh College Edition*, *Thorndike-Barnhart Advanced Dictionary*. And that's not all.

The *Oxford English Dictionary* has started work on a 3 volume supplement that will cover new words and meanings since about 1914. The first volume came out in 1972. The *Oxford English Dictionary* seems to be doing a more thorough job on computer terms than on general electronics! The entry for "control" in the new *Oxford English Dictionary* supplement quotes the 1948 MIT glossary and the 1955 glossary of the British Standards Institute. But the *Oxford English Dictionary* inexplicably skipped the electronic use of "emission" as the kind of signal (AM, FM, TV) a station sends out, although that word with that meaning has been around since at least 1927.

Almost the first thing I did when I got interested in microcomputers was to build my own glossary, starting with the lists in the back of the *IEEE Dictionary* (and acquiring other glossaries as I went). So I had a list I could check against a brand new dictionary (*American Heritage Dictionary*, College Edition of 1969) and a dictionary whose date of revision is known (*Webster's New World Dictionary*, Second College Edition of 1970).

American Heritage Dictionary started with a clean sheet. It has excellent typography but fewer entries than *Webster's New World Dictionary*. It uses larger type, and has very wide outer margins on each page where the artwork is put. Its vocabulary of computer terms includes:

accumulator, address, ALGOL, alphanumeric, analog computer;
base, bit; Boole, George; Boolean algebra;
chip, computer, computer language, converter ("a device that transforms information from one code to another");
data ("Numerical information in a form suitable for processing by computer"), *data processing, demodulation, digital computer;*
flip flop, FORTRAN;
gate ("a circuit extensively used in computers that has an output dependent on some function of its input");
hardware ("a computer and the associated physical equipment directly involved in the performance of communications or data processing functions");
information, information theory, input;
machine, machine language, memory, module, Murphy's Law;

PL/I, printer, print out (verb), print-out (noun), program (noun and verb), programmer;
readout, real time;
software, storage.

Webster's New World Dictionary has more entries and smaller type. The College Edition is its number 2 product. Some of the computer definitions are eyebrow raisers:

accumulator, address ("the location in a computer's storage compartment of an item of information, identified by a number or other code"), *alphanumeric, analog, analog computer;*
bit, Boolean Algebra;
computer, console;
data processing, decoder, digital computer;
flipflop, FORTRAN ("a digital computer language similar to algebra");
gate;
hardware;
information, information theory, input;
language ("a special set of symbols, letters, numerals, rules, etc, used for the transmission of information, as in a computer"), *logic* ("the systematized interconnection of digital switching functions, circuits, or devices, as in electronic digital computers");
machine language, memory;
printer, printout (noun), program (noun and verb, two definitions each), *programmer and programmer* (as shirttails, undefined, following program);
random access, read ("to obtain [information] from [punched cards, tape, etc]; said of a computer"), *readout, real time, routine;*
software, storage, store, symbolic logic;
throughput, track;
word, write ("to record information in a computer's memory or on a tape, etc, for use by a computer").

Webster's New World Dictionary had a contributing editor who was then the head of the Electronics Engineering Department at Carnegie-Mellon University. *American Heritage Dictionary* had no electronics specialist identified as such on its 1969 masthead.

If you know someone who is majoring in computer and minoring in linguistics, you might set him or her on this. The sooner it's wrapped up thoroughly, the easier it will be to do thoroughly. One question that should be explored in any exhaustive treatment is why certain words were admitted to the general vocabulary, and others were not, in each edition of each dictionary.

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BYTE's Bits

A Phonograph Record in a Computer Magazine?

Bob Jones of *Interface Age* magazine has informed us of an intriguing new feature included in their May issue: a flexible plastic phonograph record bound into the magazine which contains a 4 K byte BASIC compiler plus a binary loader and two memory test programs. The program, written by Robert Uiterwyk for Motorola 6800-based systems, uses the 300 bps Kansas City standard. The record's appearance marks the first time that such a technique has been used in a magazine to disseminate software. (Readers may recognize the record as the type used to promote recordings by mail).

To recover the programs, the user plays the record on his or her phonograph in the normal manner and feeds the output to an AC-30 or similar cassette interface, the same procedure used for data cassettes. Using the "tape out" feature (available on most amplifiers) is probably the easiest way to feed the signal to the interface.

Each record is good for about 100

playings; unless they are severe, scratches have no effect on accuracy. An additional benefit of this system is the elimination of tape dropout problems. Bob has promised some 8080 and Z-80 programs for future issues. Contact *Interface Age*, POB 1234, Cerritos CA 90701. ■

KIM Has a Contest

In a press release from MOS Technology Inc, Richard Simpson has announced the KIM Software Contest, open to all KIM owners and users. The prizes are:

- First prize: KIM-3 8 K memory expansion board
- Second prize: KIMROM-1 Resident Editor/Assembler ROM Set
- Third prize through tenth prize: KIMath Source Listing and User Manual.

All entries must contain program documentation and source code listing (but a hand assembled source is allowed). All entries become the property of MOS

Technology Inc and will be turned over to the KIM Users Group for possible publication.

Entries will be judged on the basis of originality and usefulness to the user community. If external hardware is required, a schematic should be provided. Complex programs taking more than 1 K bytes of memory such as high level languages, assemblers, cross assemblers, text editors, etc, will be awarded a duplicate first prize if accompanied by working source tape or cassette. All entries must be received by July 1977. Prizes will be awarded on August 1 1977. Send all entries to KIM Software Contest, MOS Technology, 950 Rittenhouse Rd, Norristown PA 19401.

This looks like an interesting opportunity for KIM enthusiasts to earn a bit of glory. ■

The Following Was Received From American Federation of Information Processing Societies (AFIPS):

The fast growing field of personal computing will be in the national spotlight this June at the 1977 National Computer Conference in the Dallas Convention Center. Reflecting the dynamic growth and promise of the personal computing field, the 1977 NCC will feature a Personal Computing Fair, a Personal Computing Exposition, two full days of program sessions, a National Club Congress, as well as special interest sessions for computer hobbyists.

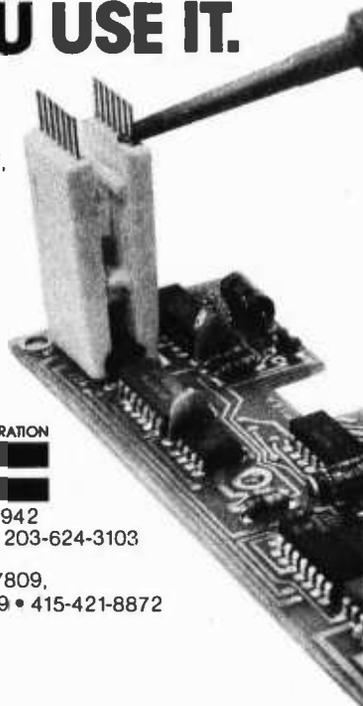
The Personal Computing Fair, scheduled to run throughout the four days of the conference, June 13 to 16, will feature operational displays and demonstrations of individually and group owned noncommercial projects. More than 100 small computing systems are expected to be displayed featuring hardware and software implementations, games, recreation, music, art, amateur radio, as well as scientific and general applications. Prizes and awards will be presented in recognition of outstanding achievement. The Personal Computing Fair will provide hobbyists with the unique opportunity to obtain new ideas for their own systems, solutions to current problems, and a wealth of "how-to" tips on personal computing.

The conference program will feature an in-depth examination of personal computing on Wednesday and Thursday, June 15 to 16. Two 3 hour panel sessions on Wednesday will examine "Personal Computing - Past, Present and Future" and "Hardware for the Computer Hobby Market." Thursday morning will feature a 3 hour panel covering "Personal Computing Software," with the afternoon devoted to the presentation of papers relating to personal computing, plus a concluding panel on "The Future of Retail Computer Stores." Each panel will feature presentations by leading authorities in the personal computing field, and will be designed to

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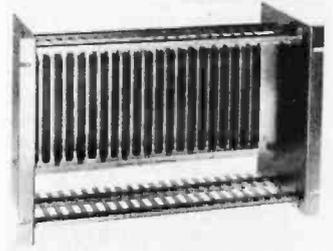
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provide attendees with the latest information on new developments, trends, and the outlook for the future. Ample time will be allotted to answering questions from those in the audience.

Plans are also under way to bring together various special interest groups in personal computing for a series of informal sessions on such topics as the building of computing kits, debugging software, use of assembly language, peripheral interfaces, cassette and disk storage, and software standards. In addition, plans are being developed for a "National Club Congress" to enable representatives of clubs from throughout the nation to exchange ideas and discuss issues relating to their activities and programs. Among expected topics will be whether or not a national personal computing association is needed, and if so, how it might be formed. Related topics are expected to include hardware/software standards, a possible national program library interchange, and the establishment of educational seminars.

In addition, the 1977 NCC will feature a commercial exhibition by equipment manufacturers and suppliers of personal computing products and services. The Personal Computing Exposition will be in the North Hall of the

Dallas Convention Center, one level below the main NCC exhibit hall.

Information on the 1977 NCC may be obtained from AFIPS Headquarters, 210 Summit Av, Montvale NJ 07645, or by calling (201) 391-9810. ■

A Calgary, Alberta Store . . .

The Computer Shop is the name of a new store which sells IMSAI, Digital Group, Interlalia, Cromemco, Lear-Seigler, Polymorphic Systems, Southwest Technical Products, Morrow, Mini-Term and TDL products to central Canadians. The shop sent us a flier, with a handwritten note that the typical prices are USA prices plus about 25%. The store is located at 3515 18th St SW, Calgary, Alberta CANADA T2T 4T9. ■

Another Dallas Area Store

KA Electronic Sales, a Dallas distributor of industrial electronic components to both businesses and individuals, has opened a computer store at 1220 Majesty Dr in the Brookhollow Industrial Pk, Dallas TX.

The KA Computer Store currently

markets central processing units and peripherals by several manufacturers including IMSAI, Southwest Technical Products Corporation, Lear-Siegler Terminals, Solid-State Music and The Digital Group.

KA also supplies electronic components and parts as an industrial and retail distributor, and has a second electronics parts walk-in store located at 1117 S Jupiter, Garland TX. ■

How to Get a BASIC Source Listing

Dr Dobb's Journal of Computer Calisthenics & Orthodontia, in its January edition of this year, has published the complete source and object code assembly listing of the Lawrence Livermore Laboratory BASIC interpreter developed by John Dickenson, Jerry Barber, John Teeter and Eugene Fisher. The interpreter is a 5 K byte program designed to be loaded in PROM or ROM. It includes a floating point arithmetic package. *Dr Dobb's* is located at People's Computer Company, POB 310, Menlo Park CA 94025. ■

Question:

Dr Chuck Adams of the Texas A & M University EE Department posed the following question in a recent phone conversation: "Who invented the D flip flop?" Can a reader supply the answer to this query, for publication in a future issue? ■

Want to Find Out Who's a Professional Computer Scientist?

The 1977 Association for Computing Machinery Roster of Members, an alphabetic and geographical cross-listing of the names and addresses of more than 35,000 ACM members as of January 1 1977 is now available.

The Roster may be ordered from the ACM Order Department, POB 12105, Church St Station, New York NY 10249. Prices are \$7 to members and \$25 to nonmembers, prepaid. ■

Guide to Buzzwords

"Sherry's Guide to Data Communications Buzzwords" is the name of a 24 page booklet of words and definitions which are commonly used in the data communications field. Write for your complementary copy, available from: Public Relations Dept, International Communications Corp, 8600 NW 41st St, Miami FL 33166. ■

Survey Sweepstakes Results . . .

In November 1976 BYTE ran a random survey of readers, to gather data for editorial and marketing purposes about this crazy field. Of approximately 2100 survey questionnaires mailed, 1448 were returned prior to the deadline of Novem-

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ber 15 1976. As an incentive to return the survey, we offered five Life Subscriptions to BYTE, commencing with the expiration of the current subscriptions of the winners. The following five individuals were drawn at random from the returned survey sweepstakes entry blanks (which were kept separate from the actual questionnaires in order to keep the questionnaires anonymous and private).

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S Charleston WV 25303

Mark T Marshall
18229 Topham St
Reseda CA 91335

Howard Rothman
86-25 Van Wyck Expy
Briarwood NY 11435

in publication of a report late this year. An international advisory group will review the report before any moves to approach Third World nations about hosting the spaceport.

Total cost of the study is estimated at \$50,000, to be provided by contributions from philanthropic, space oriented organizations and individuals.

Although the project is at an early stage, it has drawn support from leading aerospace figures. Among the advisors are Prof Freeman Dyson of the Institute for Advanced Studies, Princeton; Dr Phillip K Chapman, a former astronaut; Dr Raymond Bisplinghoff, past NASA associate director and research director; and Dr George Robinson of the Smithsonian Institution.

Three study groups will be responsible for preparation of the final report. The Government Launch Activities Committee will examine ways that governmental organizations could use an equatorial spaceport to their advantage, and the Private Users Committee will investigate opportunities for nongovernmental groups. The Freeport Design Committee, chaired by Stanford freeport specialist Dr Avlin Rabushka, will recommend possible sites and alternative legal and economic configurations of the site.

Copies of an eight page brochure

describing Earthport are available for \$.50 per copy. Suggestions about the project are welcomed by the foundation. For further information, write Mark Frazier, Space Freeport Project, Sabre Foundation, 221 W Carrillo St, Santa Barbara CA 93101.

Project Members (Partial Listing)

Advisory Board: Dr George Robinson, Smithsonian Institution; Dr Philip Chapman, former astronaut; Prof Freeman Dyson, Institute for Advanced Studies, Princeton; Pat Gunkel, Hudson Institute; Dr Larry Smarr, Harvard Astrophysics Center; Prof Alvin Rabushka, Hoover Institution, Stanford; Dr Raymond Bisplinghoff, former NASA associate director, NASA research director, and dean of engineering at MIT; Robert Prehoda, consultant.

Study Groups. Government Launch Activities Committee: Arthur M Dula JD, chairman. Private Users Committee: Robert W Poole Jr, chairman; Raymond L Kendall, program development manager, Motorola Inc; Paul Siegler, president, Earth/Space Inc. Freeport Design Committee: Prof Alvin Rabushka, chairman; Michael Bader, assistant director, NASA-Ames; Jerry Glenn, consultant.

Executive Director: Mark Frazier. ■

Technology Fact

Attention ~~Science Fiction~~ Lovers . . .
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A group of aerospace specialists has begun to investigate the prospects for a satellite launch center at the equator. Sponsored by the Sabre Foundation, the group hopes to determine the extent of interest by government and private organizations in an "Earthport" that would be open to peaceful users from every nation.

Equatorial sites offer cost savings for most satellite launches because the earth's spin gives rockets a boost into orbit. Several nations now operate equatorial launch sites of varying sizes, but none are international.

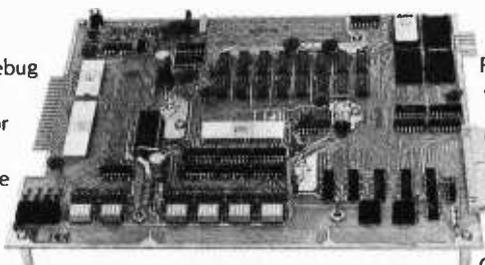
"In the past six years, aerospace companies such as Boeing and General Dynamics have explored the possibility of providing commercial launches from the equator," said the director of the study, Mark Frazier. "We plan to work with representatives of private organizations as well as governments to determine what environment would be best suited for them."

The initial stage of the study is intended to assess international interest in establishing a "space freeport," and will be completed within the next four months, according to Frazier.

Project members will then evaluate the economic, technical, legal and political aspects of establishing an international launch site, culminating

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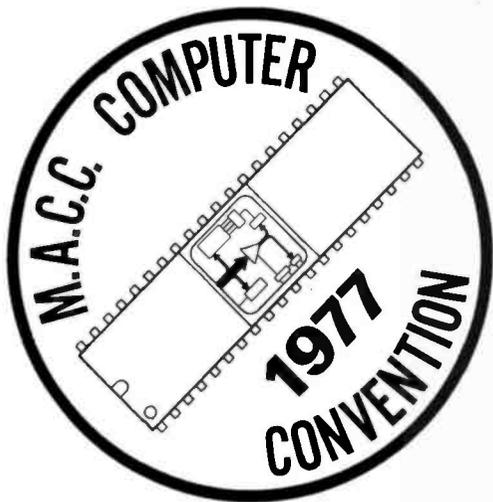
puter system elements are incorporated in this fully assembled, tested single-board computer. Some unexpected features of the HAL MCEM-8080 are: hardware "front panel" which allows setting a breakpoint and manual control of the computer; 1K ROM Monitor/Debug Software (with user callable, Intel compatible I/O routines) which greatly simplifies program development; Parallel and Serial I/O on the board; and very reasonable prices.

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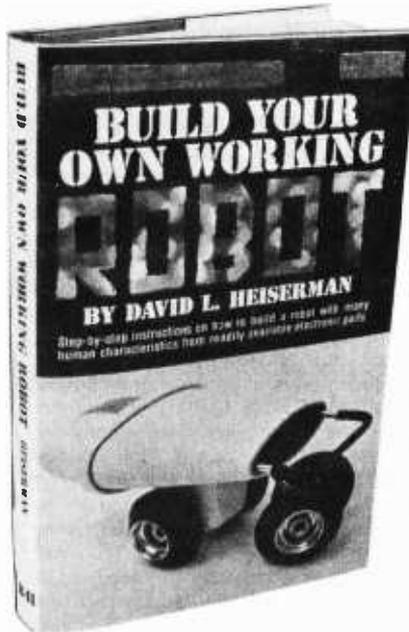
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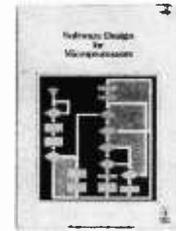


—A Dictionary of Microcomputing by Philip E Burton. In the opinion of BYTE's editor, Carl Helmers, "This is one of the best *designed* and executed dictionaries of computer related terms yet seen on the market. It is of particular relevance to those individuals who want a good general reference to numerous technical terms, broadly covering hardware and software fields as currently practiced." This new hardbound edition is part of the Garland Reference Library of Science and Technology. \$12.50.



—Scientific Analysis on the Pocket Calculator by Jon M Smith, published by John Wiley & Sons. This book is another in a set of source books for mathematical analysis using the contemporary products of technology. It is oriented to the pocket calculator, yet it will provide you with algorithms and methods useful with any personal computer which implements the scientific and analytical functions found on a good pocket calculator. For a more complete description, see the book review on page 120 of the December 1976 BYTE; or order its 392 pages of detailed technical information and review its use for yourself. \$13.75.

—Build Your Own Working Robot by David L. Heiserman, published by Tab Books. This book will not tell you how to build Robbie, the robot of Forbidden Planet, or a classical android of science fiction. What it will introduce you to is the problems of making a robot mobile device called Buster III, using pre-microprocessor TTL integrated circuits for all logic functions. It is a must book for background reading, but much of the logic can be extremely simplified using today's microprocessor technology. Use this book as a first look at these problems from which you can build further and more elaborate solutions. Softbound, \$5.95.



—Software Design for Microprocessors. This stand alone guide to microprocessors has been designed by the people at Texas Instruments to convey knowledge to the first time user of microprocessors. This excellent source book of computer concepts begins with an outline of the basic principles of the general purpose computer, its machine architecture, software, and methods of addressing. It proceeds to discuss how to build software, what is involved in documenting what you've done once you've done it, the mechanics of programming, and specific examples using the TI TMS-1000, TMS-8080, TMS-9900 and SBP0400 designs. You'll find a thick hardcover textbook filled with over 370 pages of useful information including a comprehensive glossary of microprocessor terminology, among several other detailed appendices. \$12.95.

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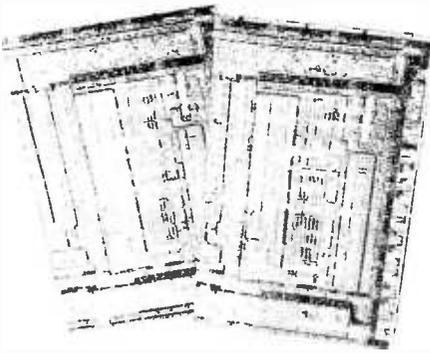
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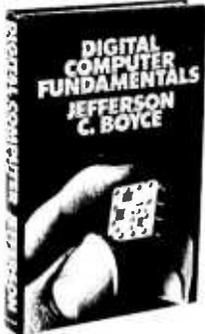


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Continued from page 52

Selectric mechanism found in the Keyboard Printer is a set of switch contacts which are closed by movement of the tilt rotate bails, and by movement of the cams in various stages of the printing cycle. These contacts can also be seen in photo 4. Again, no electric power is applied to these contacts inside the Selectric, but six of them, called C1 to C6, are wired together thru certain pins in the receptacle at the back of the machine (more on this later). For printed output, these contacts can be tested to determine when the printing cycle is complete. For keyboard input, there is another set of contacts which must be tested at the proper instant in order to capture the code for the key just depressed. Other contacts are provided which make it possible to determine whether the machine is currently locked in upper or lower case, whether the end of line margin stop has been reached, and so on. According to the documentation, the contacts are rated for 40 mA at 10 V (minimum) to 300 mA at 48 V (maximum).

BCD and Correspondence Machines

At this point, I should clear up the mystery surrounding the differences between the so-called "BCD" and "Correspondence" versions of the Selectric Keyboard Printer. There are differences in three areas:

1. The arrangement of characters on the typeball that is used.
2. The arrangement of the fingers on the interposers connected to particular keys.
3. The code obtained for keyboard input at the 50 pin receptacle when a key is pressed.

The Correspondence version is the simpler of the two. All of the office typewriters are built this way, and nearly all the typeballs available from IBM use the Correspondence arrangement of characters. In a Correspondence encoded Keyboard Printer, the tilt and rotate bail contacts are wired directly to the 50 pin receptacle, and so the code obtained when a key is pressed is the actual tilt rotate code. Note that the tilt rotate code is the same for, say, an upper case A and a lower case a, so the current state of the shift contacts must be checked whenever a character is read.

Many Selectric Keyboard Printers were built for use in equipment which employed a 6 bit byte and the old BCD (binary-coded decimal) character code, and so IBM developed the "BCD" version of the Selec-

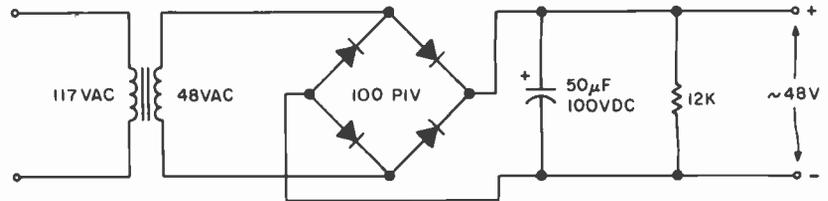
tric. In this machine, the tilt and rotate contacts (there are several sets of contacts for each bail) are wired through a maze of diodes and shift contact connections to yield a unique 6 bit code for all of the essential characters in the BCD set. Hence the code which reaches the 50 pin receptacle can be read directly into a 6 bit byte, and the shift contacts themselves need not be tested. Of course, a 6 bit byte can represent only 64 different characters, and after allowing for the digits and various special characters, there was room for only the upper case alphabets. In fact, because of the limitations of wiring through diodes and switch contacts, only 48 distinct codes are actually produced. Even so, in order to accomplish this wiring feat, it was necessary to move some of the essential characters to convenient spots on the typeball, and hence the interposers with certain finger combinations also had to be moved around in order to preserve the usual layout of the keyboard. This is why the characters are all mixed up when you type manually on a BCD machine with a Correspondence typeball. Indeed, just to make everything fit together, IBM puts only the upper case characters on most of the typeballs intended for use with the BCD machine. (An exception is the Model 963 typeball which is used in many timesharing terminals.) But, in fact, the mechanism is still capable of tilting and rotating to any character position.

What does all this mean for the computer hobbyist? If you are using the Selectric as a printer only, it makes no difference whether you have a BCD or a Correspondence machine, since in either case you have direct access to the tilt and rotate magnets. By energizing the proper combinations of the seven magnets, you can use both BCD and Correspondence typeballs with either machine. (My Selectric is a BCD machine and I regularly use it with a Correspondence encoded Courier 72 typeball.)

If you want to use the Selectric keyboard for computer input (and you want upper and lower case), or if you want to use the machine off line with a variety of Correspondence encoded typeballs, you are considerably better off with the Correspondence version of the Keyboard Printer. But, since most of the units available through surplus channels (at least at reasonable prices) are BCD machines, you may have to settle for one of these. With some mechanical and electronic skill (and lots of courage), you could convert a BCD machine into a Correspondence version by:

1. rearranging the interposers to match the Correspondence typeball arrangement.

Figure 2: A very simple power source for the unregulated DC used to power the solenoids of the Selectric Keyboard Printer.



- tearing out all the wiring for BCD code generation and replacing it with direct connections from the bail switch contacts to the 50 pin receptacle.

So much for the theory of operation of the Selectric mechanism. Now let's get on to the design of an interface unit which will let us control the Selectric printer using standard TTL level signals from a computer output port. Mindful always of our potential exposure to Murphy's Law, we will keep this interface as simpleminded as possible. Readers with more sophistication in electronics may use this approach as a jumping-off point (so to speak) for their own designs.

Interface Design

To control the operation of the Selectric printer we must provide three types of functions:

- Signal conversion of TTL levels to magnet currents.
- Code conversion of ASCII codes to tilt rotate code.
- Control and timing to type successive characters, wait for carriage return, etc.

It seemed to me that the most appropriate division of labor was to provide the first function in hardware, and the second one in software. Signal conversion requires an external power source, while code conversion requires some flexibility to accommodate different typeballs. For the third function, I have experimented with both open loop control (realized entirely in software) and closed loop control (which uses a hardware feedback signal); both approaches will be discussed briefly here.

Signal Conversion

For signal conversion, we simply need a power source for the Selectric magnets and a means of switching the power on and off using TTL level signals. For the power source, we need a maximum of about 1 A of DC (for seven simultaneously energized magnets at 125 mA per magnet) in the range of 43 to 53 V. The source need not be regulated nor even filtered. (See "Watts Inside a Power Supply," by Gary Liming, January 1977 BYTE, page 42, for a further discussion.) Figure 2 is a circuit diagram for the power supply which I built around a \$4 surplus transformer. The only really essential element is the full wave rectifier. The capacitor was included simply to jack up the voltage of the particular transformer I was using to the point where it would energize the magnets.

To switch power on and off, I used a set of reed relays (optoisolators or power transistors could be used instead). These particular reed relays have a coil resistance of 290 ohms, so they can be driven by an ordinary TTL gate (17 mA at 4.8 V, or 10 TTL loads). They are available from Digi-Key Corporation, POB 677, Thief River Falls MN 56701, for \$1.70 each (part number 5VRR). I used a total of 12 relays, six for the print magnets (since I forgot about the "check" magnet) and six for the most important control functions (space, backspace, tab, carriage return, and upper and lower case shift).

The reed relays were each connected to a computer output port and a Selectric magnet through the circuit diagram shown in

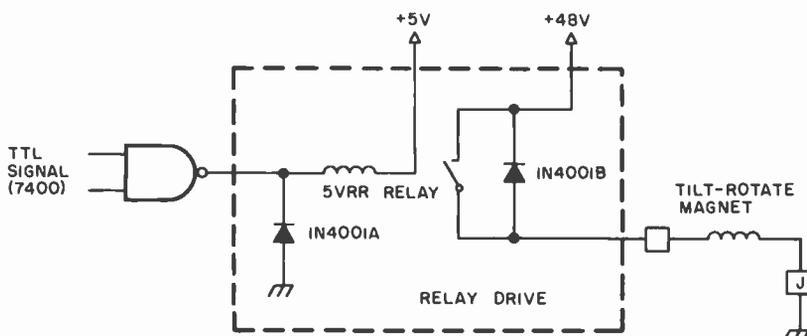


Figure 3: Switching of the solenoid actuator magnets in the Selectric Keyboard Printer is accomplished by this basic circuit. A reed relay which is within the drive capabilities of TTL is driven from a TTL logic gate, with protection against back EMF provided by the diode A. The reed relay, in turn, drives the magnet in the printer from the 48 V (nominal) supply of figure 2. Diode B provides back EMF protection for the relay contacts to prevent arcing which would shorten the life of the relay. The dotted line outlines the detailed circuit repeated many times in figure 4.

figure 3. Here the 1N4001 diodes protect the TTL gate and the reed switch from voltage transients in the two coils. Since I needed a standard TTL buffer to provide enough current for each reed relay, and since I wanted to economize on my use of output ports, I used a seventh control line to switch between the six print magnets and the six control function magnets. The resulting circuit diagram is shown in figure 4. The lettered squares which terminate the reed switch contact lines refer to pin designations on the Selectric's 50 pin receptacle (see below). Photo 5 shows the physical layout of the components of figure 4 in the interface which I built. Most of the wiring is Vector Slit n' Wrapped on the other side of the square piece of Vectorboard.

This construction layout is not recommended! Allow yourself much more room for repairing, replacing or adding components (like a seventh pair of reed relays!). A length of scrapped telephone cable makes a good connection between the interface and the Selectric itself. Also shown in photo 5 is a 50 pin connector which plugs into the

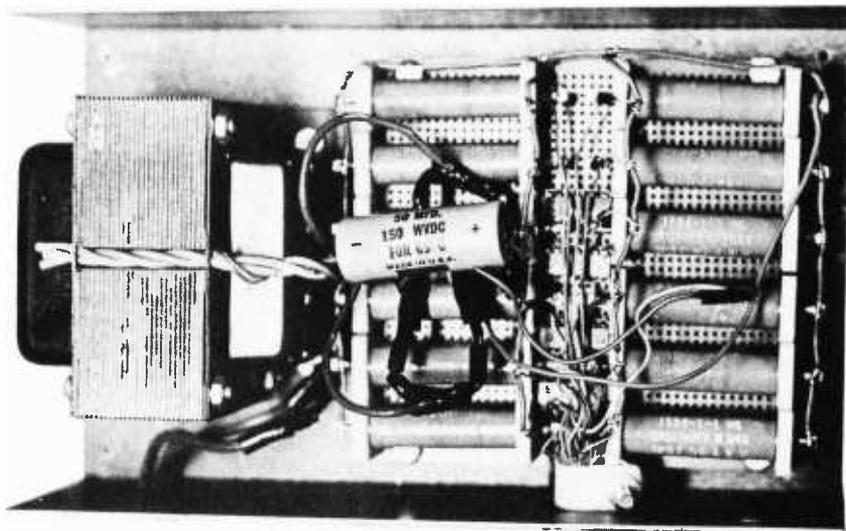


Photo 5: Physical layout of the components of the interface box which houses the circuit described in this article.

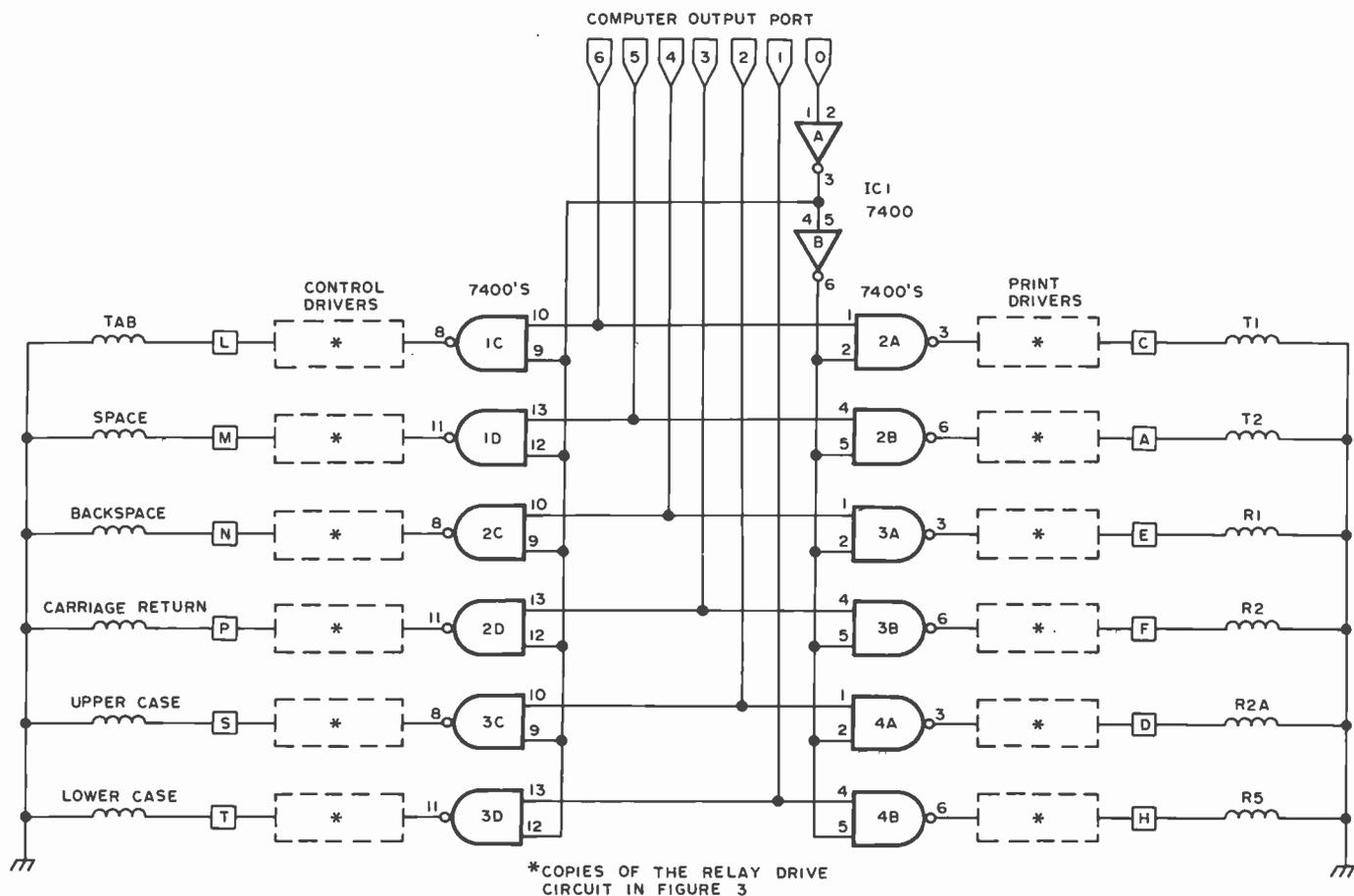


Figure 4: The complete interface schematic. The 7400 NAND gate logic is used to select either the drivers for the miscellaneous control functions, or the drivers for the print commands. The basic drive circuit of figure 3 is repeated once for each magnet in the printer.

Pin		Function
A	←	T2
B	←	Check
C	←	T1
D	←	R2A
E	←	R1
F	←	R2
H	←	R5
J	←	Magnet Common
K	←	Keyboard Lock
L	←	Tab
M	←	Space
N	←	Backspace
P	←	Carriage Return
R	←	Index
S	←	Upper Case Shift
T	←	Lower Case Shift
U	←	Red Ribbon Shift
V	←	Black Ribbon Shift
W	→	C1 N/C
X	→	Contact Common
a	→	Feedback N/C
b	→	Feedback N/O
e	→	End of Line N/C
f	→	End of Line N/O
n	→	C1 N/O
r,s,t,u,v,w	→	BCD Bit Lines

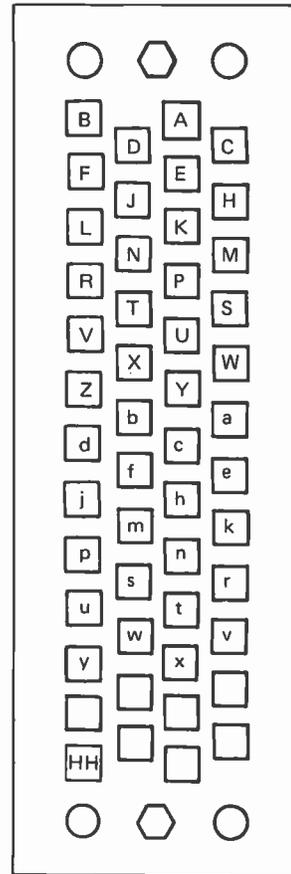


Figure 5: The Selectric Keyboard Printer receptacle pin identifications. This receptacle can be purchased as a spare part through an IBM office. The arrows in this table indicate direction of the signal: A left arrow indicates drive to the printer (typically a magnet) from a source in the interface; a right arrow indicates a sensor contact in the printer.

receptacle at the back of the Selectric, which I obtained from my local IBM branch office for \$20 (IBM part number 1167134). The more important pin designations on this connector are shown in figure 5.

Code Conversion

Assuming that the ASCII code is used for characters inside the computer, the process of code conversion is basically just a simple table lookup: The 7 bit ASCII code is used as an index into a 128 byte table to obtain the 6 bit tilt rotate code. Since the tilt rotate code for a given character may vary depending on the typeball that is used, it should be possible to switch between several 128 byte tables. This is easily done by indexing from a pointer to the base of the table as shown in figure 6.

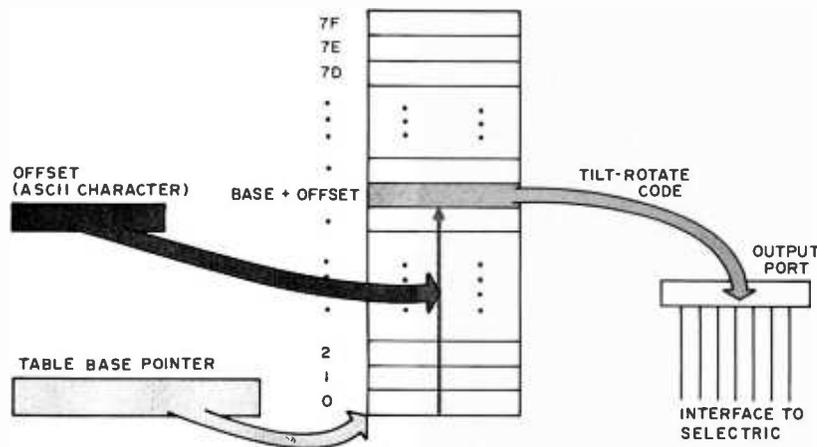


Figure 6: Table structure for the conversion of ASCII to Selectric coding. The table base pointer identifies the start of the table. There should be one table for each different ball coding scheme employed. The ASCII character value is added to the base address giving an address in the table. At this address is found the code which is sent to the output port. The logic of sending the code to the output port is given in detail by figure 8.

The main complication in code conversion is the handling of upper and lower case. At any given time the Selectric Keyboard Printer is locked into one case or the other. If the machine is locked in upper case and the next character to be printed is an upper case A, we need only send out the appropriate tilt rotate code. But if the next character is a lower case a, we must energize the lower case shift magnet, wait for the machine to shift into lower case, and then send out the tilt rotate code. This is easily accomplished by using a seventh bit in the table entry byte for each ASCII character to indicate whether it is to be printed in upper or in lower case.

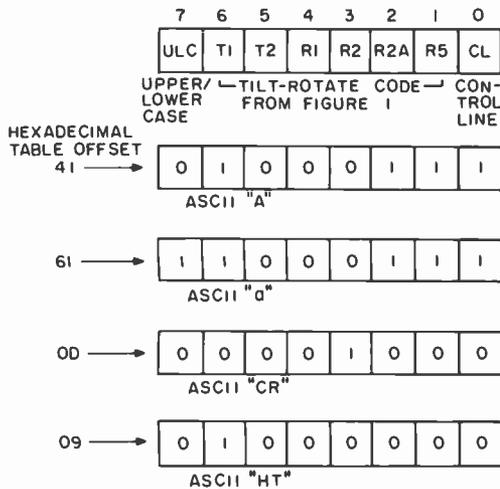
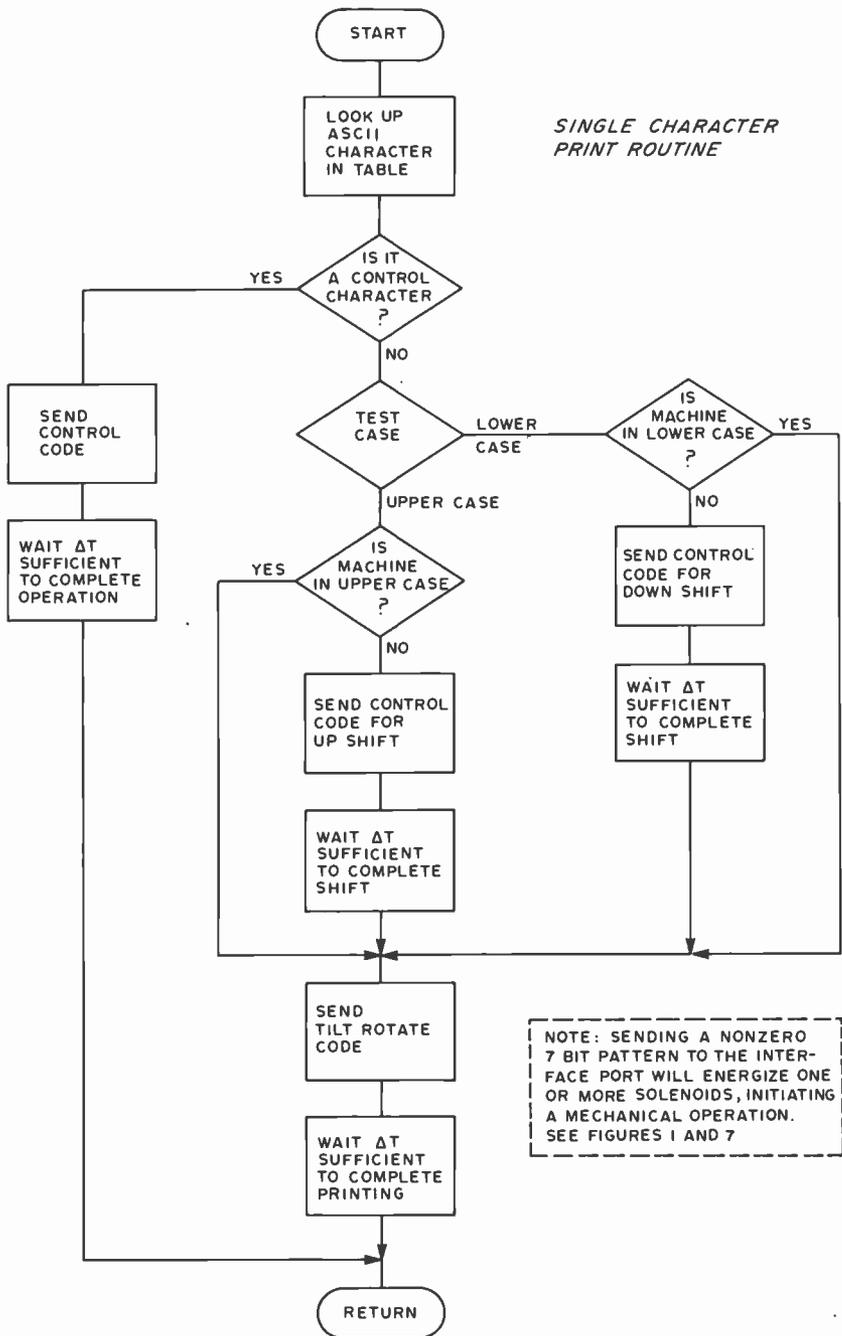


Figure 7: The coding scheme for each conversion table entry is given by the general box at the top of this diagram. Bit 7 tells the software whether the mechanism should be in the upper or lower case mode. (The need to shift explicitly in a Selectric is reminiscent of the shift requirements of Baudot Teletypes.) The tilt rotate code contained in bits 6 thru 2 is derived from figure 1 for each character in the table. (For other ball arrangements, a version of figure 1 would need to be generated.) The low order bit of the word is used to indicate to the logic of figure 4 whether a control command (0) or print command (1) is being sent.

The last problem in code conversion is the handling of control functions such as carriage return, tab, backspace, etc. Fortunately, the ASCII character set assigns unique 7 bit codes for functions such as these. For example, the ASCII carriage return character (hexadecimal code 0D) can be used for carriage return, and the ASCII horizontal tab (hexadecimal code 09) can be used for the tab function. Since in my interface a special control line determines whether the six output ports affect the print magnets or the control function magnets, I can use the eighth bit in each table entry to set the control line appropriately. The table entries for the printable characters have this bit set to 1, with six bits providing the tilt rotate code; the entries for the control characters have this bit set to 0, with the bit corresponding to the given control function magnet set to 1 and the other five bits set to 0. This encoding is illustrated in figure 7.

Once we have this encoding of the information needed for code conversion, the actual program logic to accomplish the conversion is straightforward. A flowchart of the logic is presented in figure 8, and an

Figure 8: A flowchart giving the logic of a simple open loop driver program which takes a given ASCII character, looks up its table entry, and then takes appropriate printer actions. As an open loop program, each time delay in this chart (the ΔTs) is picked to reflect the worst case response time for the action involved. This makes the Selectric type successful, but does not optimize operation for the maximum speed, since as everyone knows, the worst case is often not identical with the typical value of a parameter.



**CHARACTER OUTPUT ROUTINE FOR
SELECTRIC KEYBOARD PRINTER**

OUTCH	TAY	ASCII character to index register
	LDA (TABPT), Y	get code byte from table
	LSR A	test low order bit
	BCC CTL	0 means control character
	ROL A	test high order bit
	BMI LOWER	1 means lower case character
	LDX #4	code for upper case shift
	LDY CASE	check current case
	BEQ OK	0 means upper case
	INC CASE	indicate shift to upper case
	JMP SHIFT	go initiate shift operation
LOWER	LDX #2	code for lower case shift
	LDY CASE	check current case
	BNE OK	-1 means lower case
	DEC CASE	indicate shift to lower case
SHIFT	STX PORT	send shift code to port
	JSR ENERG	for 10 milliseconds
	LDY #60	delay for 60 milliseconds
	JSR WAIT	until shift operation is done
OK	STA PORT	send tilt rotate to port
	JSR ENERG	for 10 milliseconds
	LDY #50	delay for 50 milliseconds
	JSR WAIT	until print operation is done
	RTS	return to calling program
CTL	ROL A	restore control code
	STA PORT	send to output port
	JSR ENERG	for 10 milliseconds
	LDY #120	delay for 120 milliseconds
	JSR WAIT	until control operation is done
	RTS	return to calling program
ENERG	LDY #10	set up for 10 millisecond delay
	JSR WAIT	loop for that long
	LDY #0	send 0s to output port
	STY PORT	to turn off magnet current
	RTS	return to caller
WAIT	LDX #200	number times thru inner loop
LOOP	DEX	decrement inner loop count
	BNE LOOP	loop until count is 0
	DEY	decrement outer loop count
	BNE WAIT	loop until count is 0
	RTS	return to caller

Listing 1: 6502 assembly language source code of a program which implements the logic of the flowchart in figure 8. This program is a subroutine which will drive the Selectric Keyboard Interface in an open loop mode and is run on a KIM-1 system.

equivalent assembly language program for the MOS Technology 6502 used in my system is shown in listing 1. In this simple version of the program, delay loops are used for timing purposes, and sufficient time is allowed either to print a character or to complete the worst case control function (carriage return across the entire length of the page). Of course, this version of the program will operate the Selectric at far less than its maximum rated speed, and will monopolize the processor's time while waiting for completion of each operation. In order to improve on this, we turn next to the subject of control and timing.

Control and Timing

Now that we have a working Selectric interface, we can turn our attention to two major improvements: driving the Selectric at maximum rated speed, and minimizing use of the processor's time for Selectric control.

To drive the Selectric at full speed we can adopt an approach of "open loop" control or "closed loop" control. Open loop control

involves keeping track of the carriage position, margin, tab stops and similar information in software (changing the margin and tab stop information via software interpreted commands), and calculating the delay time necessary for each operation. Closed loop control involves testing the Keyboard Printer's switch contacts to determine when each operation has been completed. The worst case delay approach used in the program of listing 1 is a simplified version of open loop control. For full speed operation, the closed loop approach is much simpler and more reliable; so let's consider it here.

Nearly every mechanical operation opens or closes some set of switch contacts inside the Selectric. Sets of contacts are wired to the 50 pin receptacle in a variety of ways to reflect operations such as printing, tabbing, backspacing, etc. We will not consider all the possible methods of achieving feedback control using these contacts, but will outline one particularly simple approach, which remains to be tested in my own system. The pin labeled a on the receptacle is wired through a set of normally closed contacts, and the pin b through corresponding normally open contacts, associated with the set of common contacts connected to pin X. Figure 9 shows how these contacts may be debounced to yield a clean TTL level signal (ignoring the nominal voltage ratings for the contacts). Here we use the last half of the 7400 package left over from figure 4. During any printing or control function operation, pin a will go from ground to +5 V and back to ground again, while pin b does the reverse. Hence the feedback line will go from logic 1 to 0 to 1. By sensing this change in software through a loop testing the feedback input port after energizing the magnets, we can closely control the operation. When the line goes to logic 0, we can turn off current to the magnets, and when it returns to logic 1, we are ready to start the next operation.

The second problem we face in control and timing is how to minimize use of the processor's time for Selectric control. Here, of course, is where the interrupt system comes into play. If we are using the circuit outlined in figure 9 for closed loop control, we can tie the feedback line to a processor interrupt rather than to a data input port. If we are relying instead on open loop control, we can use a programmable interval timer which is capable of causing an interrupt as an alternative to delay loops. The software to handle interrupts from the Selectric is slightly complicated by the need to shift between upper and lower case prior to typing the next character, but this can be handled by initiating the shift operation and

then arranging to retry the character printing operation on the next interrupt, at which time the Selectric will be locked into the proper case.

Actual Experience

Hopefully this article has given the reader all the information he or she needs to build a Selectric Keyboard Printer interface similar to, or better than mine. Lest you are unduly emboldened by the foregoing discussion, however, consider what can go wrong.

I carefully tested the interface in stages, by using an ohmmeter to verify that bit patterns sent to my computer output port closed the proper combinations of reed switches, and by testing the power supply on some of the Selectric's magnet coil connections. Nevertheless, when I first tested the entire setup, I thought I saw a blue flash around one of the reed relays when I tried to pulse the R2 magnet. Nothing seemed to happen when I tried again, except that the R2 magnet wasn't being energized. Then, listening carefully, I heard a telltale sizzling sound that sent me leaping for the electric outlet. The R2 reed relay had stuck closed, and on further examination I found that most of the arc suppressing diodes inside the Selectric had been destroyed. After painstakingly replacing the R2 reed relay and installing the diodes visible in photo 5, I tried again. This time I found out why the reed relay, like its replacement, was sticking closed! The R2 magnet in the unit I purchased had been burned out and was a short circuit. No wonder the unit was a surplus item.

Not willing to give up, I managed to remove the coil from the R2 magnet core, and replace it with the coil from the unused (by me!) check magnet. After this feat, I found that when I typed manually on the keyboard, only @s, Os, and a few other characters could be printed! Only after hours of reading and experimentation did I discover that the adjustment of the plate holding the magnet armatures in place (which I had removed to change the coils) was critical, and could be set only by considerable trial and error.

These are the kinds of things that can go wrong. You cannot be too careful in playing with these machines! Readers certainly should investigate the possibility of an IBM maintenance contract on at least the mechanical portion of the Keyboard Printer, which need not be too expensive.

And, to conclude, although I probably never would have undertaken this project had I known at the outset what it would ultimately entail, it certainly is satisfying to have that Selectric typing away under the

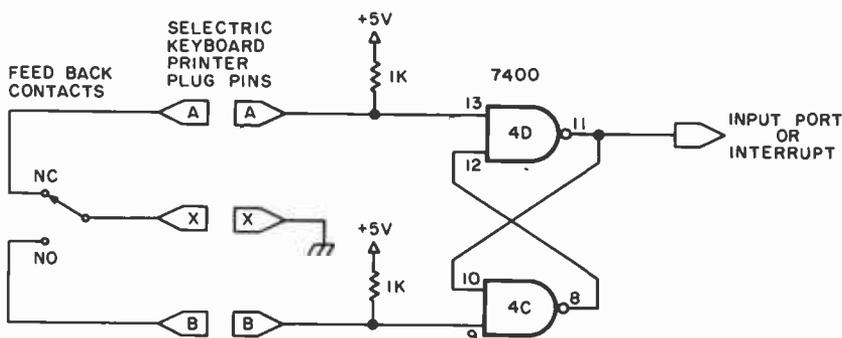


Figure 9: A circuit for debouncing the feedback information generated by contacts in the printer which are mechanically linked to the action. Using the feedback pulse to drive an input port or interrupt line can result in operation at the maximum possible speed since the timing is now on an "each case" basis rather than "worst case."

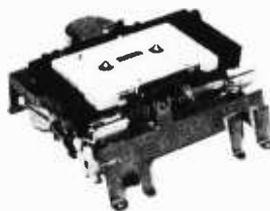
control of my home computer. To anyone else who is ready to undertake such a project, I hope that this article has helped, and I wish you the best of luck. ■

BIBLIOGRAPHY

"IBM Selectric Input-Output Writer: An Exciting Advance in the Field of Input-Output Media," Form # 543-0033-1. This manual is absolutely essential since it gives circuit diagrams, timing charts, and end views of the magnets and switch contacts.

"IBM Selectric I/O Keyboard Printer: Customer Engineering Manual of Instruction," Form # 241-5159-3. This or a similar manual is very valuable for understanding the mechanical functioning of the Keyboard Printer.

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A 6800 Selectric IO Printer Program

Listing 1: The listing of the Selectric printer interface routine for a 6800 system driving the IO version of the standard office typewriter. This listing is extracted from two assemblies done using the Southwest Technical Products Corporation's version of the M6800 self assembler. The first part of the listing is the actual code, and the second part is a table of Selectric correspondence codes which is referenced using ASCII codes as an index into the table which is computed at CONVI.

```

SWTPC M-6800 ASSEMBLER
ENTER PASS : 1P,1S,2P,2L,2T

00001          NAM      SELECTRIC
00002          *SELECTRIC DRIVER PROGRAM FOR SWTPC 6800 ASSEMBLER
00003          OPT      S
00004          OPT      O
00005          OPT      L
00006 0100     ORG      $0100
00007 0100 7E 17CD  JMP      START
00008 0212     ORG      $0212
00009 0212 BD 17F4  JSR      START1  CALL OUTPLT(NEW)
00010 08F7     ORG      $08F7
00011 08F7 18FF  FDB      $18FF  MAKE ROOM FOR PATCH
00012 093D     ORG      $093D
00013 093D 1900  FDB      $1900
00014 13D6     ORG      $13D6
00015 13D6 1900  FDB      $1900
00016 17CD     ORG      $17CD
00017 17CD C6 FF  START LDA B  #$FF  INITIALIZE PIA
00018 17CF F7 8000 STA B  PIAOUT
00019 17D2 7F 8002 CLR  PIACHK
00020 17D5 C6 04  LDA B  #$04
00021 17D7 F7 8001 STA B  PIAOUT+1
00022 17DA F7 8003 STA B  PIACHK+1
00023 17DD C6 81  LDA B  #$81  START ALWAYS IN LOWER CASE
00024 17DF F7 8000 STA B  PIAOUT
00025 17E2 F7 18CA STA B  CASE
00026 17E5 FE 181A LDX  COUNT1  SETUP TNER FOR SHIPT CYCLE
00027 17E8 FF 18CB STX  COUNTR
00028 17EB BD 18C3 JSR  TIMER
00029 17EE 7F 8000 CLR  PIAOUT
00030 17F1 7E 0300 JMP  $300  GOTO MAIN PRGM
00031 17F4 84 7F  START1 AND A  #$7F  RESET PARITY
00032 17F6 FF 18CD STX  SAVEX  SAVE XREG FOR MAIN PRGM
00033 17F9 81 10  CMP  A  #$10  TRAP HOME-UP
00034 17FB 26 02  BNE  CR
00035 17FD 20 04  BRA  CR1  PRINT IT AS CR,LF
00036 17FF 81 0D  CR  CMP  A  #$0D  TRAP CR
00037 1801 26 0B  BNE  SP
00038 1803 86 84  CR1 LDA  A  #$84
00039 1805 FE 181C LDX  COUNT2  SETUP TIMER FOR CR,LF
00040 1808 FF 18CB STX  COUNTR
00041 180B 7E 18BA JMP  EX1
00042 180E 81 20  SP  CMP  A  #$20  TRAP SPACE
00043 1810 26 05  BNE  CONVO
00044 1812 86 88  LDA  A  #$88
00045 1814 7E 18B4 JMP  EX2  GO PRINT BUT DO NOT RESET MSB
00046 1817 7E 1880 CONVO JMP  CONVRT
00047 181A 2000  COUNT1 FDB  $2000
00048 181C 1000  COUNT2 FDB  $1000
00049 181E 0400  COUNT3 FDB  $0400
00050 1880     ORG      $1880
00051 1880 81 20  CONVRT CMP A  #$20  IS IT A PRINTING CHARACTER?
00052 1882 22 02  BHI  CONV1  YES
00053 1884 20 39  BRA  EXIT  NO
00054 1886 B7 18D0 CONV1 STA A  TABLEP+1  CONVERT CODE

```

The following letter and listing 1 were received from an Italian reader of BYTE, Fulvio Guzzon of Rome. Fulvio purchased the same print mechanism (IBM Model 735 IO typewriter) which is described by Dan Fylstra in his article in this issue. We're treating Fulvio's letter as a short article, since its technical content is far above that of the usual letter. The listings photographically reproduced here were typed on pin feed paper using his printer mechanism. The text of his letter was submitted using a text editor with the Selectric IO mechanism as its output.

I understand there is some interest among your readers in using a Selectric typewriter for hard copy. As you can see I have funneled an editor program (SWTPC) and an assembler program (SWTPC, too) through a Selectric typewriter. [The original of this note was typed on the Selectric.] I bought the machine on the surplus market in Boston and it had some problems: It was stuck in upper case by a bolt screwed on the right side of the frame, it had some unrecoverable backlash in the head rotate mechanism, and many feedback and interlock contacts were missing or badly damaged. I had the machine serviced here in Rome (Italy) and at last, with a new carriage, a new motor (here we have 220 V 50 Hz power), and a new set of shift magnets, the printer was ready. I decided to use it only as a printer in order to reduce the hardware and software effort to a minimum.

On the underside of the machine there are seven printing magnets. In table 1 I have paired them with the bits from 0 to 6. Seven transistors provide for the interface between the PIA and the printer.

There are seven more magnets for the machine commands: space, backspace, tab, carriage return, index (line feed), upper case, and lower case; so seven more transistors are required. Seven output lines from the PIA in slot 0 are switched between the two sets of magnets by digital logic. The various feedback and interlock contacts were wired in series and filtered for bounce by a condenser and a software loop. The conversion table shown in the assembly listing provides for the characters used on the so called "Correspondence" balls. As I later found out, there are minor variations between the balls of this series.

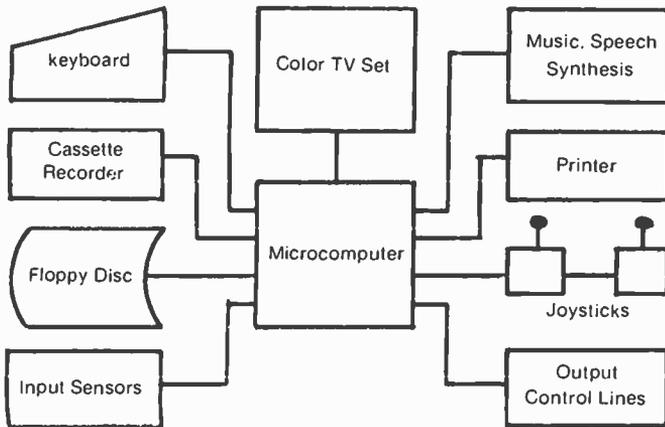
The MSB in the table is set when the character to be printed is on the upper case half of the ball. (The upper or lower case of ASCII code bears no relation to the upper or lower side of Selectric golf balls). The MSB of the output byte to the printer

Listing 1, continued:

00055	1889	FE 18CF	LDX	TABLEP	
00056	188C	A6 00	LDA A	0,X	
00057	188E	26 02	BNE	CASECK	IS IT AVAILABLE SOMEWHERE ON
00058				* THE BALL?	
00059	1890	20 2D	BRA	EXIT	NO,RETURN
00060	1892	2A 04	CASECK BPL	CASELW	MSB CLEAR?
00061	1894	C6 C0	LDA B	#\$C0	NO,CHECK IF PRINTER IS IN UC
00062	1896	20 02	BRA	SKIP	
00063	1898	C6 81	CASELW LDA B	#\$81	YES,CHECK IF PRINTER IS IN LC
00064	189A	F1 18CA	SKIP	CMP B	CASE NEW CHAR. SAME HALFBALL
00065				*AS THE PREVIOUS ONE?	
00066	189D	27 13	BEQ	PRINT1	YES GO AND PRINT IT
00067	189F	F7 8000	STA B	PIAOUT	NO, ROTATE BALL 180 DEGREES
00068	18A2	F7 18CA	STA B	CASE	AND RECORD IT
00069	18A5	FE 181A	LDX	COUNT1	SETUP TIMER FOR SHIFT CYCLE
00070	18A8	FF 18CB	STX	COUNTR	
00071	18AB	8D 16	BSR	TIMER	
00072	18AD	7F 8000	CLR	PIAOUT	
00073	18B0	8D 11	BSR	TIMER	
00074	18B2	84 7F	PRINT1 AND A	#\$7F	RESET CASE BIT AND
00075	18B4	FE 181E EX2	LDX	COUNT3	SETUP TIMER FOR PRINT CYCLE
00076	18B7	FF 18CB	STX	COUNTR	
00077	18BA	B7 8000 EX1	STA A	PIAOUT	NOW PRINT
00078	18BD	8D 12	BSR	WAIT1	
00079	18BF	FE 18CD EXIT	LDX	SAVEX	RESTORE X REG
00080	18C2	39	RTS		GO AND FETCH NEXT CHARACTER
00081	18C3	FE 18CB	TIMER LDX	COUNTR	
00082	18C6	09	LOOP	DEX	
00083	18C7	26 FD	BNE	LOOP	
00084	18C9	39	RTS		
00085		8000	PIAOUT EQU	\$8000	
00086		8002	PIACHK EQU	\$8002	
00087	18CA	00	CASE	FCB	
00088	18CB	0000	COUNTR	FDB	
00089	18CD	0000	SAVEX	FDB	
00090	18CF	1800	TABLEP	FDB	\$1800
00091	18D1	8D F0	WAIT1	BSR	TIMER
00092	18D3	C6 01	LDA B	#1	SETUP MASK
00093	18D5	F5 8002	BIT B	PIACHK	PRINT CYCLE STARTED?
00094	18D8	27 F7	BEQ	WAIT1	NO
00095	18DA	7F 8000	CLR	PIAOUT	YES ON IT'S WAY
00096	18DD	8D E4	WAIT2	BSR	TIMER
00097	18DF	F5 8002	BIT B	PIACHK	READY FOR A NEW ONE?
00098	18E2	26 F9	BNE	WAIT2	NO
00099	18E4	39	RTS		YES!
00100	18E8		ORG	\$18E8	SHIFT START OF SYMBOL TABLE
00101				*TO NEXT PAGE	
00102	18E8	C1	FCB	#\$C1	
00103	18E9	20	FCC	5,	
	18EA	20			
	18EB	20			
	18EC	20			
	18ED	20			
00104	18EE	FFFF	FDB	\$\$\$\$	
00105	18F0	C2	FCB	#\$C2	
00106	18F1	20	FCC	5,	
	18F2	20			
	18F3	20			
	18F4	20			
	18F5	20			
00107	18F6	FFFF	FDB	\$\$\$\$	

When you get your home or office computer, will you know what to do with it?

The typical home or small business computer system starts with a microcomputer, keyboard, cassette recorder, and TV set. From there you can add the peripherals, sensors, controllers, and other devices you need for your own special applications.



Creative Computing Magazine is dedicated to describing applications for home, school, and small business computers completely and pragmatically in non-technical language. You won't need a Ph.D in Computer Science, or a technical reference library, or a computer technician beside you to get these applications up and running. We give you complete hardware and software details. Typically, applications utilize commercially available systems. However, if an application needs a piece of home-brew hardware, we tell you how to build it. Or if it requires a combination of high-level and machine language code, we give you the entire listings along with the flowcharts and algorithms.

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8. Fuel economizing systems
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3. Loan/annuity/interest calculations and analysis
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Circle 87 on inquiry card.

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SYSTEM 5000

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SYSTEM 5000 is the programmable clock kit that makes kit-building a new experience. The system has been designed to meet a variety of particular requirements and tastes, and programming techniques are used to create a truly individualized timepiece. Numerous functions and features are provided for maximum flexibility and adaptability, and any or all can be used to construct many different types of time keeping and timing devices.

SYSTEM 5000 is not a simple LED time of day clock, but a full feature digital timing system. Programming is accomplished by connecting the appropriate jumpers and switches to produce the desired system configuration. Complete assembly and programming manuals are included.

SYSTEM 5000 has a fluorescent readout panel with four 0.5" numerals that brighten and dim automatically according to the ambient light. This unique digital display provides optimum readability at all times from almost any viewing angle.

SYSTEM 5000 can be built as a desk clock, alarm clock, calendar clock, or all of these in one full-feature timepiece. The Duplicate Time Register can monitor elapsed time or another Time Zone such as GMT. A ten minute "ID" reminder capability is included for Radio Station use. A quartz time base is available for high precision, stability and uninterrupted operation if the AC line should fail.

SYSTEM 5000 can automatically control AC or DC accessories up



to 700 Watts by adding the optional relay. Plug in your radio or stereo to construct a full-function clock radio that puts you to sleep with gentle music and wakes you to music, a tone, or both. The system will also control TV's, small appliances, or other accessories. SYSTEM 5000 can be used to construct timers for a variety of applications. It is ideal for automatic process timers and controllers in laboratories, workshops, and engineering facilities. SYSTEM 5000 includes all components, speaker, two time setting switches, and comprehensive instruction and programming manuals. Case & switches for programming additional functions are not included but available as options \$29.95

FEATURES AND SPECIFICATIONS

Timekeeping Functions	Display	General
<ul style="list-style-type: none"> • Time of Day Register • Duplicate Time Register • True 24 Hour Alarm • Duplicate 24 Hour Alarm • 10 Minute Snooze on Alarms • True Four Year Calendar • One Hour Down Counter 	<ul style="list-style-type: none"> • Bright 4-Digit Fluorescent Panel • Automatic Brightness Circuit • 12 or 24 Hour Display Format • PFI and Power Failure Indication • 1 Hr Activity Indicator • Power-On Clear • Direct Drive Eliminates all RFI 	<ul style="list-style-type: none"> • Forward or Reverse Time Setting • Reset and Count Inhibit Controls • Seconds Display • Single 9 Volt Battery Backup • 700 Watt Relay Optional • 60 or 60 Hz, 117Vac, 3 Watts • 1.5" x 4"W. x 4"D.

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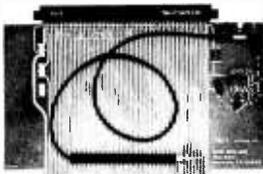
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A Full Size Floppy Disk with Altair Interface



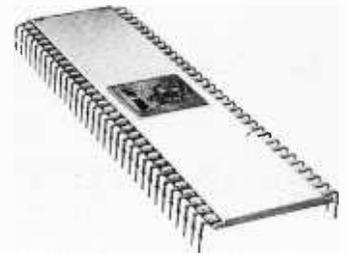
Peripheral Vision, POB 6267, Denver CO 80206, has announced this full size floppy disk for the Altair bus. Prices start at \$750 for the interface card kit and one assembled and tested drive. A 24 V at 2 A power supply is also available in kit form for \$45 or assembled for \$65; and a cabinet is offered for an additional \$85. The Peripheral Vision floppy disk interface card supports eight drives and according to the press release, stores over 300,000 bytes per floppy. A bootstrap EROM is included to make system start-up automatic.

The floppy is completely Altair bus compatible, and interface cabling is included. The Peripheral Vision floppy disk drive itself is manufactured by Innovex, and comes assembled and tested. A disk operating system with file management system is included on a floppy disk cartridge.

Circle 615 on inquiry card.

New Technology for the 9900 Family

Texas Instruments Inc has introduced a new version of the 9900 architecture in the form of this SBP 9900, which uses integrated injection logic (called I²L in much of the engineering and design literature). The key features of this new product are in the subtleties of using the



chip at a hardware level (software is identical to the previous TMS 9900 and Texas Instruments' 990 line of mini-computers which use this architecture). For the hardware designer or homebrew hacker, this new premium version of the design gives a wide (-55°C to +125°C) operational temperature range, infinitely variable clock rates from DC to 3 MHz (power consumption levels vary with the rate) and TTL compatibility with wide tolerances on power supply voltages. This is, however, a premium device intended for applications which require ruggedness. The ceramic packaged SBP 9900 starts at a \$386 price in 100 piece quantities. Reader inquires should be directed to Texas Instruments Inc, Inquiry Answering Service, POB 5012, M/S 308 (attn: SBP 9900), Dallas TX 75222. ■

Circle 616 on inquiry card.

A Rugged Z-80 Product



Cromemco has sent along this press release picture of the new Z-2 processor which is the latest product of their laboratory. This processor uses their Z-80 processor board which according to the company is available in a fast 4 MHz version. Coupled with Cromemco's other peripherals which also work at this speed, in principle this one has one of the fastest processors yet available in a commercial product. The basic box contains the processor card, a mother board with 21 card slots for Altair compatible cards, and a heavy duty power supply intended to suffice for all system needs including floppy disk drives. Naturally, in addition to the \$595 kit version of this processor, you'll want to have some dedicated monitor ROM and extra peripherals, but the price makes it an attractive way to start building a system. Cromemco has software of a monitor, assembler, and a BASIC with processor control extensions. The firm also makes numerous peripherals including digital to analog interface cards, ROM memory cards and a color graphic product. Cromemco is located at 2432 Charleston Rd, Mountain View CA 94043. ■

Circle 617 on inquiry card.

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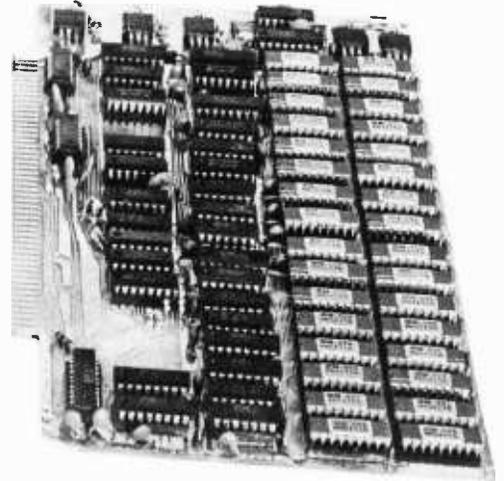
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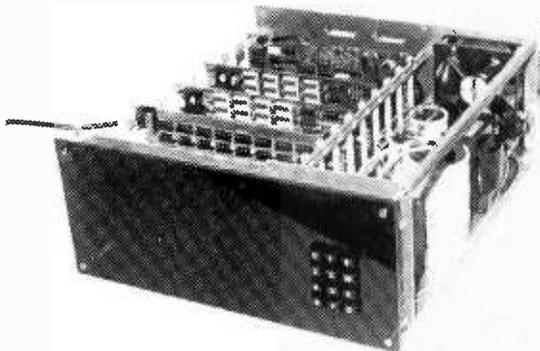
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Real, Palo Alto CA 94306, sends this picture of a new Altair compatible, assembled and tested 16 K byte memory board with a price of \$485 or \$30 per installed 1024 bytes. The board contains its own refresh control logic and uses 4 K dynamic memory chips with a total board power consumption of 5 W. External to the board, this product looks like a static memory board and has no wait states since the refresh is transparent. The number of personal computers with a saturated 64 K memory address space is likely to go up as more and more products such as this reach the market. ■

Circle 618 on inquiry card.

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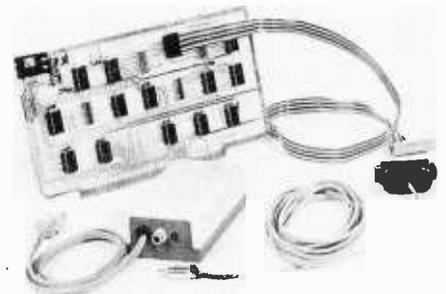
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Circle 260 on inquiry card.

146

Controlling Those Necessary Bells, Whistles and Other Goodies

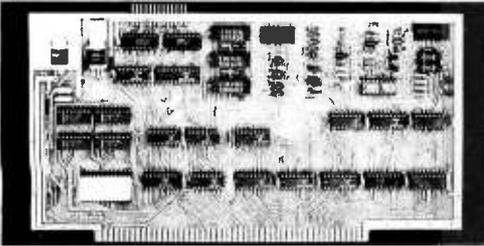


Comptek has designed and manufactured this interesting array of components with the needs of the real world interface in mind. The heart of the control setup is of course a typical micro-computer, an Altair bus compatible machine into which the main "control logic interface" board plugs. Out of the board comes a ribbon cable to a DB-25 connector at the back edge of the cabinet. The DB-25 is in turn the recipient of a cable which goes to your remote power unit, the box shown in this picture. This box, which is one of many which may be driven from one control logic interface board, contains the optically isolated 400 W controller for 110 VAC. The costs of this interface in kit form are \$189 for a 16 channel

PC3216 control logic interface card, plus \$39.50 for each PC3202 400 W power control unit kit. Other options include fully assembled versions at higher prices and a 32 channel PC3232 control logic interface card. A product like this is needed when you want to have your computer drive 110 VAC appliances, lighting circuits and other household electrical loads. Comptek is located at POB 516, La Canada CA 91011. ■

Circle 619 on inquiry card.

Morrow Tape Interface



Morrow's Micro Stuff, POB 6194, Albany CA 94706, has introduced this Altair bus plug-in unit which generates and reads data on up to three channels of audio recording in the Kansas City standard of recording, at 300 bps. Also thrown into the board is a serial port to allow communication with a Teletype with reader control, as well as any RS-232 serial device. Also included is an 8 bit parallel board for use with parallel interfaced peripherals such as keyboards or tape readers. A ROM on the board holds 512 bytes of programming for the cassette interface, UART simulation, and transfer to or from your 8080's memory or the 512 byte programmable memory region on the board. This board is available in kit form for \$120 or assembled and tested at \$165 with warranty. The product is marketed by mail or through computer stores. ■

Circle 620 on inquiry card.

Matrox Video Display with External Sync Capability

One fascinating possibility for the use of small computers is in combination with standard video signals for various purposes. (An example might be display of product feature data along with digital messages in a merchandising situation such as a department store.)



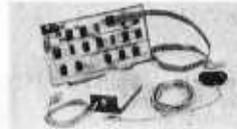
comptek Boards DO Something



CL2400
Real Time Clock

\$98—Kit \$135—Assembled

If your system needs to know what time it is, our CL2400 is the board for you. The present time in hours, minutes, and seconds is always available for input, and is continuously updated by the highly accurate 60 Hz power line frequency. Need periodic interrupts? The CL2400 can do that, too, at any of 6 rates. Reference manual with BASIC and assembly language software examples included.



PC3200
Power Control System

PC3232 \$299—Kit \$360—Assm.
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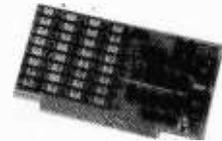
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Matrox Electronic Systems, POB 56, Ahuntsic Sta, Montreal, Quebec CANADA H3L 3N5, has just announced the new member of its video RAM family of devices. This is the MTX-1632SL, shown here in front of a television driven by a mixed signal generated from another video source and the character generator. Since the horizontal and vertical synchronization can be slaved to the external source, such as the standard television video picture in this case, it is possible to combine the displays. The application of this \$225 module (lower prices in larger quantities) provides ample opportunities for the imagination. ■

Circle 621 on inquiry card.

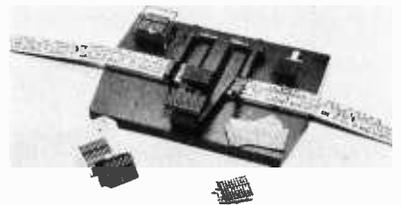
Ruggedized Power Supplies



Calex Manufacturing Company Inc, 3305 Vincent Rd, Pleasant Hill CA 94523, has come out with this sealed power supply black box which is designed for use with small dedicated microprocessor systems using the 8080. The supply transforms an AC input at 110 V (or several other voltages) into +12 V at 225 mA, -5 V at 20 mA, and +5 V at 1.25 A. These voltages are sufficient to run a typical 8080 system with 8080, 8024, 8228, two 2708s and several programmable memory parts with about 500 mA of TTL logic power left over on the +5 V lines. ■

Circle 622 on inquiry card.

A Splice in Time Saves Nine?



Master Digital Corporation, 1308-F Logan Av, Costa Mesa CA 92626, has come out with this \$60 tape splicing jig which can be had with various options for tapes from 5 to 8 channel widths. The company also makes pressure sensitive mylar tape patches in lengths from one to 12 inches, opaque or clear. ■

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Personal Computing Expo is being produced by H.A. Bruno & Associates, Inc., a firm in the exposition and promotion fields since 1923. Highly skilled in the production and promotion of consumer and trade shows, the company currently promotes the American Energy Expo, the National Boat Show, Auto Expo/New York. Promotion assistance also is currently rendered to the National Computer Conference and the Triennial IFIPS Congress in Toronto.

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PCE PERSONAL COMPUTING EXPO • NEW YORK COLISEUM
OCTOBER 28, 29, 30, 1977

SR-52 Card BLACKJACK

Michael J Garvey, vice president and systems consultant with General Computer Services Corporation, 2308 Central Av, Middletown OH 45042, sends in a game program for the SR-52, which was accompanied by this note:

I appreciate your article in your December issue on the "buried gold" in

Location	Code	Key	Commentary	
000	01	1		
001	51	SBR	Deal card to dealer	
002	32	SIN		
003	42	STO		
004	00	0	Store as face down card.	
005	03	3		
006	02	2		
007	51	SBR	Deal card to player.	
008	32	SIN		
009	01	1		
010	51	SBR	Deal card to dealer.	
011	32	SIN		
012	41	GTO		
013	12	CB	Jump around subroutine	
014	46	*LBL		
015	32	SIN	Begin "Deal Card" subroutine	
016	42	STO		
017	00	0		
018	00	0		
019	22	INV		
020	57	*fix		
021	43	RCL		
022	09	9	Generate random number in R99	
023	09	9		
024	40	*X ²		
025	52	EE		
026	00	0		
027	09	9		
028	22	INV		
029	52	EE		
030	42	STO		
031	09	9		
032	09	9		
033	75	-		
034	53	(
035	42	STO		
036	55	o/o		
037	01	1		
038	03	3		
039	75	-		
040	93	.		
041	05	5		
042	54)		
043	57	*fix	Convert random number to a card value.	
044	00	0		
045	52	EE		
046	22	INV		
047	52	EE		
048	65	X		
049	01	1		
050	03	3		
051	85	+		
052	02	2		
053	75	-		
054	42	STO	Store card value	
055	00	0		
056	05	5		
057	01	1		
058	01	1		
059	95	=		Is card = 11 [ACE]? If so then go to COS
060	90	*if z		
061	33	COS		
062	22	INV		Is card in range 2 to 10? If so then go to TAN.
063	80	*if p		
064	34	TAN		
065	01	1		
066	00	0	Force card value of 10 if outside legitimate range	
067	41	GTO		
068	65	X		
069	46	*LBL		
070	33	COS		
071	36	*IND		
072	43	RCL		
073	00	0		
074	00	0	Blackjack Logic:	
075	75	-		
076	01	1	Would hand go over 2, if ace counted as 11?	
077	01	1		
078	95	=	If so then ACE = 1 else ACE = 11	
079	22	INV		
080	80	*if p		
081	34	TAN		
082	01	1		

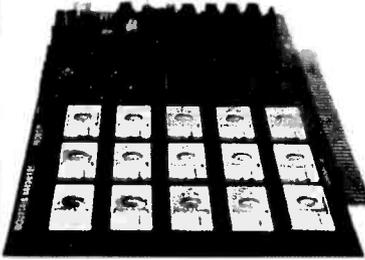
Operating Procedure For "SR-52 Card BLACKJACK":

1. Load the program card, both sides, after it has been prepared.
2. Prime the random number generator with a 9 or 10 digit number as its seed. A good choice of a priming number is the current time of day (24 hour clock), followed by the date. This gives 10 digits total in the format.
Enter: hhmmyymmdd
Then press: STO, 9, 9
3. Enter the amount of your bet and start the game:
Enter: bet amount
Then press: A
4. Outcome:
If the display flashes, then the game is over. The display shows the total of dealer and player hands.
Press: CE to stop flashing.
Then press: RCL, 9, 8 to read out the cumulative score if desired.
Go to Step 3 to restart game.
If the display is not flashing, the dealer's face down card is not shown.
5. Player options:
To take a "hit,"
Press B
Then go to step 4.
To "stand,"
Press C
Then go to step 4.
To "double down" (if player's hand is 10 or 11 and not initial deal).
Press D
Then go to step 4.

Notes on operation of the game:

- The player always wins at 21.
- House always draws to 16 or lower, stands on 17 or higher.
- On a "push" (both hands equal) neither side wins unless player has 21.

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Dates for the Camps are June 19-24, June 26-July, July 10-15, July 17-22. The fee, including tuition, room and board, is \$125.

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For further information contact Dr. John Kinney, Rose-Hulman Institute of Technology, 5500 Wabash Ave., Terre Haute, Indiana 47803.

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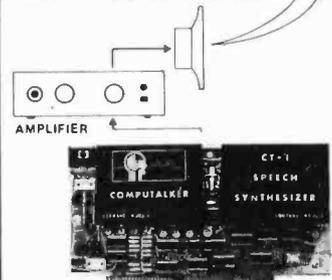
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Listing 1, continued:

Location	Code	Key	Commentary
083	46	*LBL	Store adjusted card value.
084	65	X	
085	42	STO	
086	00	0	
087	05	5	
088	46	*LBL	Add card to total hand of receiver.
089	34	TAN	
090	43	RCL	
091	00	0	
092	05	5	
093	36	*IND	Game entry (start here)
094	44	SUM	
095	00	0	
096	00	0	
097	56	*rtn	
098	46	*LBL	Clear bet in R19
099	11	A	
100	47	*CMs	
101	42	STO	
102	01	1	
103	09	9	Go to 000
104	86	*rset	
105	46	*LBL	
106	14	D	
107	02	2	
108	49	*PROD	Multiply bet by 2
109	01	1	
110	09	9	
111	50	*st fl	
112	00	0	
113	46	*LBL	Turn on "stand" flag
114	12	B	
115	02	2	
116	51	SBR	
117	32	SIN	
118	41	GTO	Deal card to player
119	45	y ^x	
120	46	*LBL	
121	13	C	
122	50	*st fl	
123	00	0	Skip around "stand"
124	46	*LBL	
125	45	y ^x	
126	43	RCL	
127	00	0	
128	02	2	From step 012
129	75	—	
130	02	2	
131	01	1	
132	95	=	
133	90	*if z	Turn on "stand" flag
134	88	*2'	
135	80	*if p	
136	87	*1'	
137	43	RCL	
138	00	0	If player's total = 21 then go to 2' else if player's total >21 then go to 1'
139	01	1	
140	75	—	
141	02	2	
142	01	1	
143	95	=	If dealer's total = 21 then go to 1' else if dealer's total >21 then go to 2'
144	90	*if z	
145	87	*1'	
146	80	*if p	
147	88	*2'	
148	22	INV	If "stand" switch on then go to 8' [to display]
149	60	*if fl	
150	00	0	
151	68	*8'	
152	43	RCL	
153	00	0	Is dealer's total >17?
154	01	1	
155	75	—	
156	01	1	
157	07	7	
158	95	=	If so then go to 3'
159	80	*if p	
160	89	*3'	
161	01	1	
162	51	SBR	
163	32	SIN	Draw another card for dealer

the Texas Instruments SR-52 Programmable Calculator; it confirmed my suspicions that my SR-52 had more power than the instruction manual said.

Enclosed is a program listing for a program that I have written that will allow you to play BLACKJACK with an SR-52. This program was the toughest that I have written for that machine, since the 224 program steps allowed just didn't seem enough for the game; several days were spent in working and reworking the code in order to get the game to fit with the features I wanted. As you can see, it just fits, exactly.

I have sent this program to you in case any of your readers would be interested in it. My family and friends have had a great deal of fun with it, and it's especially great for killing time on a long trip; one person can "stake" the "house," while another person is the player. The program automatically keeps score for the player, and even handles the "double-down" feature of the game.

I submit it for the entertainment of anyone who wants to use it.

Listing 1 shows the program code, which we typeset using column headings from the original form, along with the register allocations. The procedures for using the SR-52 Card BLACKJACK program are summarized in the box labelled "Operating Procedure."

Location	Code	Key	Commentary
164	41	GTO	Go to C
165	13	C	
166	46	*LBL	
167	89	*3'	
168	43	RCL	
169	00	0	Is dealer's total = player's total?
170	01	1	
171	75	—	
172	43	RCL	
173	00	0	
174	02	2	If so then go to 7'
175	95	=	
176	90	*if z	
177	67	*7'	
178	22	INV	
179	80	*if p	Is dealer's hand less than player's
180	88	*2'	
181	46	*LBL	
182	87	*1'	
183	01	1	
184	94	+/-	If so go to 2'
185	49	*PROD	
186	01	1	
187	09	9	
188	46	*LBL	
189	88	*2'	Otherwise, dealer wins
190	43	RCL	
191	01	1	
192	09	9	
193	44	SUM	
194	09	9	So make bet amount negative.
195	08	8	
			Add current bet to player's total score

Listing 1, continued:

Location Code	Key	Commentary
196	46	*LBL
197	67	*7'
198	43	RCL
199	00	0
200	03	3
201	85	+
202	85	+
203	46	*LBL
204	68	*8'
205	57	*fix
206	02	2
207	43	RCL
208	00	0
209	01	1
210	75	-
211	43	RCL
212	00	0
213	03	3
214	85	+
215	43	RCL
216	00	0
217	02	2
218	55	o/o
219	01	1
220	00	0
221	00	0
222	95	=
223	81	HLT

Game over, so display end game

Display dealer hand as integer, player as decimal, and await next round.



The SR-52.

Register Utilization:

ALLOCATIONS FOR SR-52 CARD BLACKJACK:

User accessible labels:	Flag Usage:
A = Start game	0 = "stand" flag
B = Hit	
C = Stand	
D = Double	

- 00 = pointer for subroutine parameter
- 01 = dealer count
- 02 = player count
- 03 = facedown
- 05 = current card
- 98 = winnings
- 99 = random number output
- 19 = current bet

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A Guide to Baudot Machines:

Part 3, A Teleprinter Test Circuit

Michael S McNatt
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Now that you've found out what type of Baudot teleprinters are available on the surplus market, and where to go to get information on how to interface them to your microprocessor, it might be convenient to build a test box to check the working condition of your new acquisition. What follows is a circuit for just such a test box which can be used to provide the 60 mA current loop required by the Baudot machine. Circuits are also included in the box to generate signals which can verify correct machine function. Although not as handy as the test box, a Baudot keyboard may be used to test a page printer of the same speed. The test box has variable control of data rate for testing all Baudot teleprinters. The test box circuits supply the following functions:

1. Standard RY test signals, either continuously or at two second intervals, for mechanical alignment purposes. These signals result in a maximum amount of slipping and sliding of adjacent gears and parts within the machine, conditions which are also most likely to cause malfunctions to

surface. The two second interval used with the RY test prevents a rather large waste of paper or tape during a long test.

2. Individual Baudot characters at two second intervals. Five bit-switches select the particular character desired (see table 1 in part 1 of this article for bit codes). One use of this function is to check various machines to see which keys are actually installed for selected bit combinations. For example, there are at least three different "figures" code assignments. Another use is in troubleshooting, for instance when the wrong character is appearing on the printer or punch in response to a processor output command. The test box can be quickly switched from "CPU" to local "char" mode to verify that the source of the problem is or is not the machine itself.

Also included in the test box is circuitry to accept serial TTL or CMOS level Baudot coded output from the processor or hardware code converter. The "MODE" switch

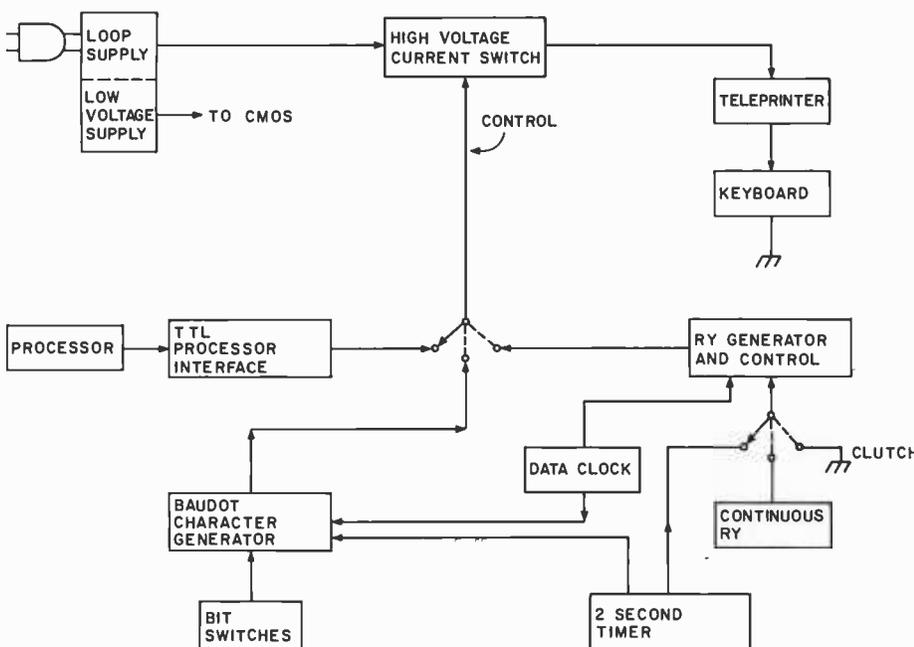
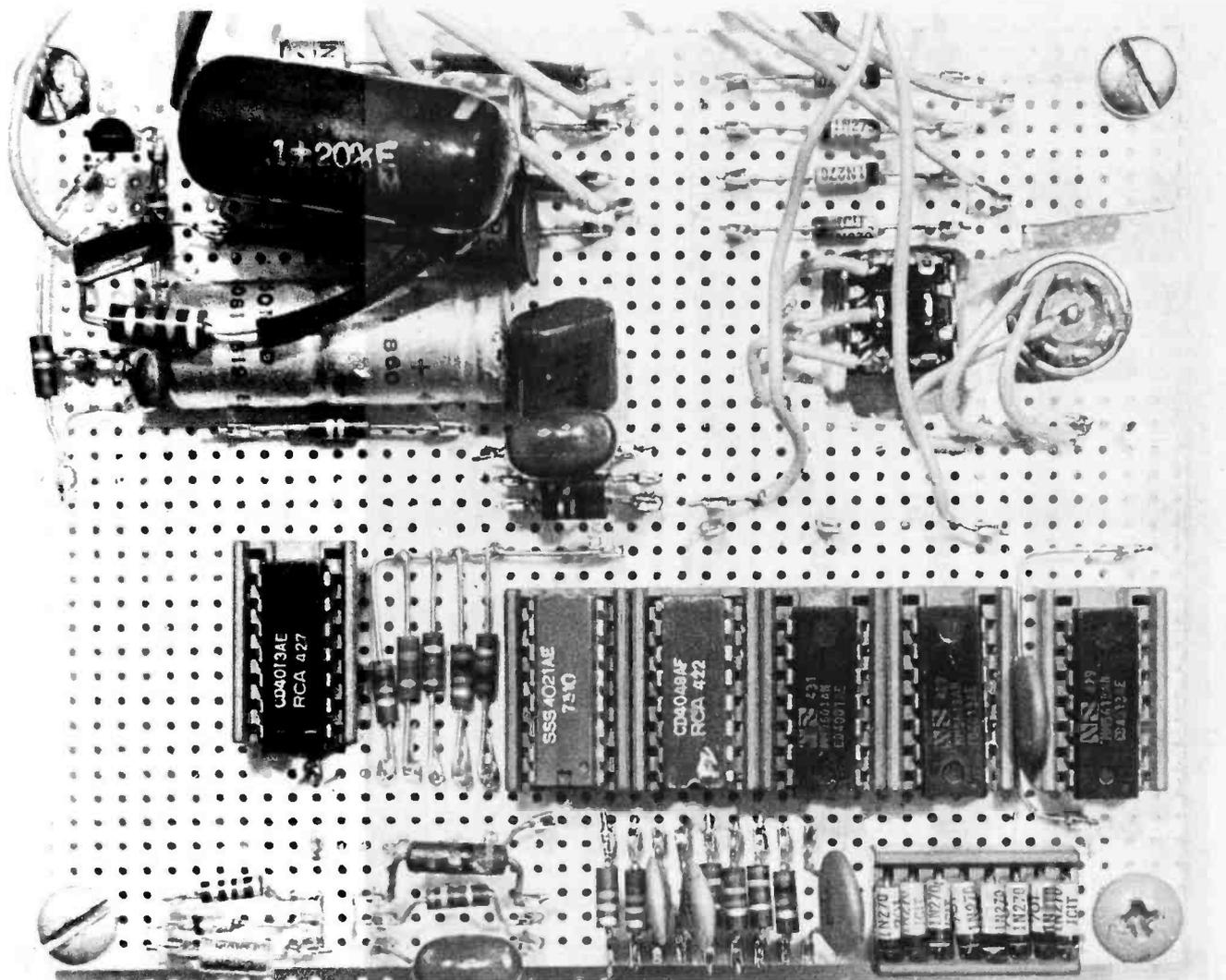


Figure 1: Block diagram of a Baudot machine test box allowing the generation of continuous or intermittent 'RY' test signals, or individual Baudot characters.



selects between processor input, RY test signals, and the individual character generator.

The absolute minimum amount of hardware one can get by with when setting up a Baudot machine is a high voltage loop supply. Some machines, Model 28s and Kleinschmidts, for example, come equipped with this supply; unfortunately, my Model 15 RO page printer didn't. It may be a good idea to borrow or build a test box anyway. When trying to determine the operating condition of a prospective Baudot machine, taking along a simple loop supply can enable at least a very minimal check of the printer mechanism. When purchasing my 15, the seller, a ham, used his own supply to demonstrate successful clutch action, then opened and closed the loop by rubbing the connections, causing random characters to be printed. Obviously, a better check would have been with a test box or a speed compatible Baudot keyboard. However, the

demonstration, in this case, was totally satisfactory because of the price, \$30!

Theory of Operation

A complete schematic of the Baudot machine test box is shown in figure 2. Many of the parts used were selected solely because they were immediately available from the proverbial junk box. (I think there are companies which would envy some of the so-called "junk boxes" I've seen belonging to hams and computer phreaks...) For instance, the transformer shown was the only one usable that came out of a thorough purging of not only my junk box but that of a fellow designer. Actually any transformer secondary from 110 V to 300 VAC may be used, as long as adequate series power resistors are included to result in a short circuit current of 65 to 70 mA. Telegraph loops, such as Western Union, usually work at 260 V. Use of the higher voltages, around 200 VDC, although requiring more power to

Photo 1: A sample layout of the Baudot test box. Note the eight diodes plugged into the 16 pin IC socket at the lower right of the picture. These diodes are being used in replacement of IC7 in the circuit of figure 2.

Photo courtesy of Don Clum

←

Figure 2: A Baudot machine test box and interface. This circuit generates 5 level Baudot code to be used in testing the working condition of a Baudot teleprinter. All resistances are measured in ohms, and unless otherwise marked are 1/4W and ±10% accuracy. The circuit allows either continuous RY test strings or intermittent 2 second test strings. The RY test string causes the most movement of gears and will show any difficulties in operation quicker than other test string combinations. All unused inputs of any CMOS integrated circuit must be grounded.

be wasted in the series resistors, results in a faster response in the printer selector magnets, which is especially important at the higher speeds such as 100 words per minute. Do *not* put the standard transient suppression diode across the coils because selector magnet release will be slowed down.

Since CMOS logic elements are used, the low voltage secondary winding can be of any voltage which will result in from 5 to 12 VDC for the +V supply. The series resistor capacitor network across high voltage transistor Q1 is used to suppress the voltage transient occurring in the magnet coils when Q1 is turned off. A low logic state at IC4, pin 10, results in a mark or clutch state, since this will switch Q2 and thus Q1 on, causing loop current to flow. Resistor R1 tied to +V insures that the machine will remain in the mark state while changing the MODE switch.

The data clock and the 2 second interval clock, IC3 and IC4, are standard CMOS oscillators. In substituting CMOS, try to avoid 4049 inverters in these oscillators. The character generator is formed by 8 bit shift register IC5, which is parallel loaded with the 5 Baudot data bits, a start bit, pin 15, and a stop bit, pin 7. The remaining bit, pin 1, is tied high. Characters are generated every 2 seconds when the load control, pin 9, is pulled high for one half period of the data clock. Flip flop IC6 synchronizes the load pulses from the slow interval clock with the much faster data clock. Note that serial input IC5, pin 11, is tied high. This provides marking 1s to be introduced into the shift register after the parallel loaded Baudot bits have been shifted out.

The RY generator is essentially a CMOS version of an RTL circuit published in *Ham Radio* magazine, March 1971, pages 23 to 29. Briefly, IC1 and IC2 form a four bit, labeled A through D on schematic, binary counter which skips two states when counting from 0 to 15. The RY pattern thus generated is shown in the timing diagram of figure 2. Referring to the schematic, counter state 1, DCBA = 0001, causes flip flop A to be reset, thus clocking flip flop B. This occurs when IC1, pin 2, goes high, causing the reset pulse at IC3, pin 4. Counter state 14, DCBA = 1110, causes flip flop A to

be set, with no clocking of flip flop B. This occurs when IC1, pin 1, goes high, causing the set pulse at IC3, pin 10.

The RY control is a three position SPDT minitoggle switch with the center position off. A neat feature of this circuit is that no matter when the RY control is switched from one position to another, complete RYs are always generated.

Miscellaneous Notes

- I used six 1N914 diodes and two resistors in place of IC7, hence the 16 pin socket was used as a diode holder, (see photo 1). The 4075s are more convenient, but the diodes were in the junk box at the time. If diodes are used, connect three of the anodes in place of IC7, pins 1, 2, 8, and the cathodes in place of IC7, pin 9, and tie a 1M resistor to ground at this point. Ditto to replace the second gate.
- To avoid the self-smoke or reach for the sky mode, all unused CMOS inputs *must* be grounded or tied to +V.
- Obviously the test box can be as simple or complex as one desires. The minimum configuration would probably be two Eveready No. 416 67.5 V batteries and a resistor in series, using the "sparking" method to generate random characters!
- Seriously, the RY generator is mandatory, with the character generator and 2 second timer as options. Of course a more sophisticated Baudot test box would have an 8080A, a 1702A, an 8251, and maybe some hexadecimal character readouts . . .

Summary

An old surplus Baudot code teleprinter is the most inexpensive hardcopy peripheral available to the computer hobbyist today. This series of articles has presented approaches to acquiring, using and testing these units, as well as sources of reference material and interfacing hardware and software. It is hoped that the information provided will greatly ease the acquisition and interfacing tasks facing the hobbyist who owns or is planning to own one of these practically indestructible machines. ■

What's New?

GNAT 3M Drive

This new GNAT MC-200 data storage system will make an interesting option for many users. This is an RS-232 interfaced serial device which plugs into a standard DB-25 connector often used



with RS-232. It gives the user a serial storage capability on 3M DCD-100 cartridges. Optional parallel IO is also available.

The unit has sufficient intelligence built in to recognize various command sequences including start write, start read, stop write, stop read, rewind, etc. All these sequences are duplicated by hardware switches on the front panel for manual control as well. The data rate for the device, which looks logically like any RS-232 serial terminal, can be set by a switch at rates from 75 to 9600 bps. Options available include parallel inter-

The Ultimate in Terminal Printers



The ultimate in terminal printing mechanisms is that provided by the Diablo HyType II printers and similar high-speed, high quality impact mech-

face, file search capability, dual drives, etc. The price of this system is \$1930, and its mix of features should prove quite useful for those who want a file storage system with the minimum of interface complexity.

GNAT is located at 7895 Convoy Ct, Unit 6, San Diego CA 92111, and delivery is quoted as from stock to 60 days ARO.■

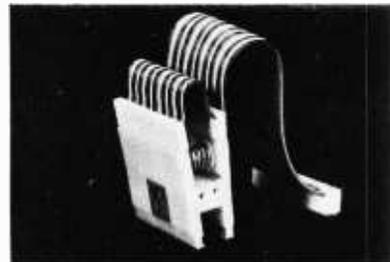
Circle 613 on inquiry card.

An Interesting Test Fixture

AP Products, POB 110, 72 Corwin Dr, Painesville OH 44077, has introduced this interesting new fixture for use in testing integrated circuit assemblies. Called the "Logical Connection," this is a preassembled ribbon

anisms. These mechanisms give print quality (with carbon ribbons) good enough to photocopy and reproduce, yet at character printing rates of up to 45 characters per second. Applied Computer Systems, 248 Sobrante Way, Sunnyvale CA 94086, has taken the Diablo mechanism and placed it into this attractive package along with a microprocessor and numerous options. The result is a hard copy oriented microcomputer with prices starting at \$4500 and options including floppy disks, memory to 64 K, plotting, sort and merge capability, down loading of programs over communications lines to large systems, and user programmability of the built-in microprocessor. Communication speeds of 600, 1200, 2400 and 4800 bps are supported, and the standard memory size is 4 K bytes. Keyboards customized for APL, ASCII and IBM 2741 compatibility are available.■

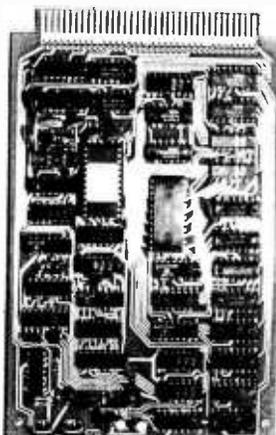
Circle 612 on inquiry card.



cable attached to an IC test clip at one end, and a flat cable socket connector at the other. The length of the ribbon cable can be chosen by the purchaser; the example here is of course rather short to illustrate the idea.■

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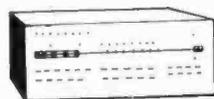
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**An Interesting New Product
for People Wanting Complete Systems**



Frank Laczko, president of TLF, POB 2298, Littleton CO 80161, called us early in February to announce his new brainchild, the Data 12 computer, shown here. This machine is an example of a completely integrated system oriented to a terminal user. Its architecture is that of the Digital Equipment

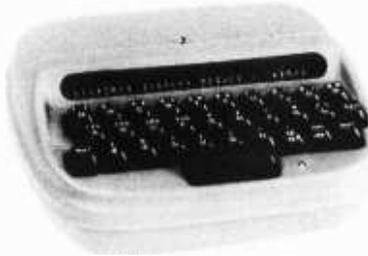
Corporation's PDP-8E, using the Intersil IM6100 microprocessor.

The \$1695 price tag of this processor with its built-in tape drive gets a PDP-8 like computer with 4096 12 bit words of user memory, serial terminal interface, tape controller with one drive built in, and a tape operating system that includes both an unattended batch mode of operation and real time task scheduling capability. The random access tape cassette drive uses a preformatted digital cassette that has an average access time quoted by the TLF press release as less than 25 seconds, with bidirectional search speeds of 100 inches per second. The tape holds a maximum of 262,144 words in 128 word blocks. The software supplied with the system is completely

oriented to an interactive keyboard, and it includes an "invisible" system executive that handles all IO scheduling, buffering and vectoring. The operating system, modeled after DEC's OS-8, is written with an eye towards device independence in the same fashion as larger computers. The system software manages named files for mass storage control, and also includes the usual text editor for program preparation, a symbolic assembler, disassembler and loaders. The system is marketed with a BASIC compiler which includes multi-segment program linkage conventions, large multidimensional arrays, string handling, and multiline user defined functions. User memory can be expanded to 32 K words, and additional peripherals are also available. ■

Circle 611 on inquiry card.

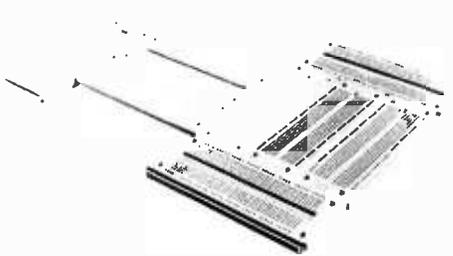
A Portable Display Terminal



The Micon KDM/1 is manufactured by Micon Industries, 252 Oak St, Oakland CA 94607. The terminal is a self-contained keyboard with 32 character alphanumeric LED display, available for \$400 mail order. The product is available in eight different colors. It is also available in many fine computer stores. Options include acoustic coupler, digital cassette tape storage and additional memory capacity to 1920 characters. ■

Circle 610 on inquiry card.

**A New Variation on Solderless
Prototyping Boards**



Continental Specialties, 44 Kendall St, POB 1942, New Haven CT 06509, has announced an interesting new variation on the solderless breadboard concept: modular units with a set of interlocking edges so that they can be built up into multiboard arrays that are rigidly held together. ■

Circle 609 on inquiry card.

BYTE's Bugs

**Oops . . . Some Phi-Deck Updates
to Freeman's Article**

I have read an article published in the March 1977 issue of BYTE. The article, "Cassette Transports for the 'Roll Your Own' Hobbyist," had some errors in it concerning the Phi-Deck cassette transports. These errors are apparently due to the normal time lags involved between the occurrence of a development and the time consumed to communicate the event.

Areas I would like to bring to your attention are:

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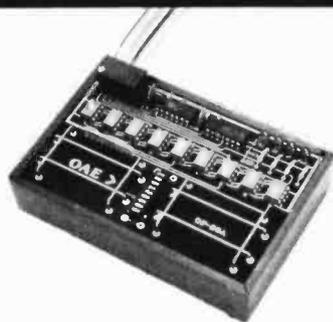
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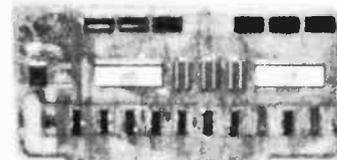
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Pictured above is the new OP-80A High Speed Paper Tape Reader from OAE. This unit has no moving parts, will read punched tape as fast as you can pull it through (0-5,000 c.p.s.), and costs only \$74.50 KIT, \$95.00 ASSEMBLED & TESTED. It includes a precision optical sensor array, high speed data buffers, and all required handshake logic to interface with any uP parallel I/O port.

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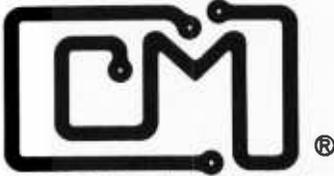
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Phi-Deck Models 1 and 2 are now built with the die cast chassis. This and other factors have increased the price of these models to \$124 in quantities of 1 to 9. The price is higher, but you can now get a precise die cast chassis for under the \$169 mentioned.

If you could update your audience about the above corrections to the article, it would be greatly appreciated.

Jack Morrow
Applications Engineering Technician
Triple I
POB 25308
Oklahoma City OK 73125

My fault. In editing, I should have correlated later Phi-Deck literature with the article, which was written in the Summer of 1976 . . . CH ■

PAPERBYTES

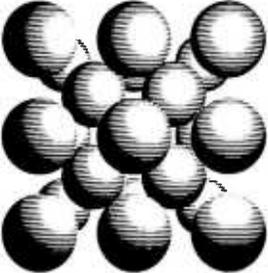
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Originally described in the April and May 1977 BYTE, PAPERBYTES is now offering Jack Emmerichs' Tiny Assembler 6800. This book contains the complete Tiny Assembler source listing plus object code in cross assembly format (space restrictions prevented printing of this material in BYTE). A bar code version of Tiny Assembler is included for convenience, as well as reprints of Jack's two articles and additional user manual materials. Tiny Assembler will run on any machine with MIKBUG and 4K of memory starting at address 0000, and is an excellent tool for the interactive development of functional blocks for a large structured program. Add it to your 6800 system and you'll have a valuable programming aid which can free you from the drudgery of machine language. The best part is the price: only \$7. Order yours today!

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5"	.98	3.42	6.34
5 1/2"	1.02	3.63	6.75
6"	1.06	3.83	7.16
6 1/2"	1.10	4.04	7.57
7"	1.14	4.25	7.98
7 1/2"	1.18	4.46	8.39
8"	1.22	4.67	8.80
8 1/2"	1.26	4.88	9.21
9"	1.30	5.09	9.62
9 1/2"	1.34	5.30	10.03
10"	1.38	5.51	10.44
10 1/2"	1.42	5.72	10.85
11"	1.46	5.93	11.26
11 1/2"	1.50	6.14	11.67
12"	1.54	6.35	12.08
12 1/2"	1.58	6.56	12.49
13"	1.62	6.77	12.90
13 1/2"	1.66	6.98	13.31
14"	1.70	7.19	13.72
14 1/2"	1.74	7.40	14.13
15"	1.78	7.61	14.54
15 1/2"	1.82	7.82	14.95
16"	1.86	8.03	15.36
16 1/2"	1.90	8.24	15.77
17"	1.94	8.45	16.18
17 1/2"	1.98	8.66	16.59
18"	2.02	8.87	17.00
18 1/2"	2.06	9.08	17.41
19"	2.10	9.29	17.82
19 1/2"	2.14	9.50	18.23
20"	2.18	9.71	18.64
20 1/2"	2.22	9.92	19.05
21"	2.26	10.13	19.46
21 1/2"	2.30	10.34	19.87
22"	2.34	10.55	20.28
22 1/2"	2.38	10.76	20.69
23"	2.42	10.97	21.10
23 1/2"	2.46	11.18	21.51
24"	2.50	11.39	21.92
24 1/2"	2.54	11.60	22.33
25"	2.58	11.81	22.74
25 1/2"	2.62	12.02	23.15
26"	2.66	12.23	23.56
26 1/2"	2.70	12.44	23.97
27"	2.74	12.65	24.38
27 1/2"	2.78	12.86	24.79
28"	2.82	13.07	25.20
28 1/2"	2.86	13.28	25.61
29"	2.90	13.49	26.02
29 1/2"	2.94	13.70	26.43
30"	2.98	13.91	26.84
30 1/2"	3.02	14.12	27.25
31"	3.06	14.33	27.66
31 1/2"	3.10	14.54	28.07
32"	3.14	14.75	28.48
32 1/2"	3.18	14.96	28.89
33"	3.22	15.17	29.30
33 1/2"	3.26	15.38	29.71
34"	3.30	15.59	30.12
34 1/2"	3.34	15.80	30.53
35"	3.38	16.01	30.94
35 1/2"	3.42	16.22	31.35
36"	3.46	16.43	31.76
36 1/2"	3.50	16.64	32.17
37"	3.54	16.85	32.58
37 1/2"	3.58	17.06	32.99
38"	3.62	17.27	33.40
38 1/2"	3.66	17.48	33.81
39"	3.70	17.69	34.22
39 1/2"	3.74	17.90	34.63
40"	3.78	18.11	35.04
40 1/2"	3.82	18.32	35.45
41"	3.86	18.53	35.86
41 1/2"	3.90	18.74	36.27
42"	3.94	18.95	36.68
42 1/2"	3.98	19.16	37.09
43"	4.02	19.37	37.50
43 1/2"	4.06	19.58	37.91
44"	4.10	19.79	38.32
44 1/2"	4.14	20.00	38.73
45"	4.18	20.21	39.14
45 1/2"	4.22	20.42	39.55
46"	4.26	20.63	39.96
46 1/2"	4.30	20.84	40.37
47"	4.34	21.05	40.78
47 1/2"	4.38	21.26	41.19
48"	4.42	21.47	41.60
48 1/2"	4.46	21.68	42.01
49"	4.50	21.89	42.42
49 1/2"	4.54	22.10	42.83
50"	4.58	22.31	43.24
50 1/2"	4.62	22.52	43.65
51"	4.66	22.73	44.06
51 1/2"	4.70	22.94	44.47
52"	4.74	23.15	44.88
52 1/2"	4.78	23.36	45.29
53"	4.82	23.57	45.70
53 1/2"	4.86	23.78	46.11
54"	4.90	23.99	46.52
54 1/2"	4.94	24.20	46.93
55"	4.98	24.41	47.34
55 1/2"	5.02	24.62	47.75
56"	5.06	24.83	48.16
56 1/2"	5.10	25.04	48.57
57"	5.14	25.25	48.98
57 1/2"	5.18	25.46	49.39
58"	5.22	25.67	49.80
58 1/2"	5.26	25.88	50.21
59"	5.30	26.09	50.62
59 1/2"	5.34	26.30	51.03
60"	5.38	26.51	51.44
60 1/2"	5.42	26.72	51.85
61"	5.46	26.93	52.26
61 1/2"	5.50	27.14	52.67
62"	5.54	27.35	53.08
62 1/2"	5.58	27.56	53.49
63"	5.62	27.77	53.90
63 1/2"	5.66	27.98	54.31
64"	5.70	28.19	54.72
64 1/2"	5.74	28.40	55.13
65"	5.78	28.61	55.54
65 1/2"	5.82	28.82	55.95
66"	5.86	29.03	56.36
66 1/2"	5.90	29.24	56.77
67"	5.94	29.45	57.18
67 1/2"	5.98	29.66	57.59
68"	6.02	29.87	58.00
68 1/2"	6.06	30.08	58.41
69"	6.10	30.29	58.82
69 1/2"	6.14	30.50	59.23
70"	6.18	30.71	59.64
70 1/2"	6.22	30.92	60.05
71"	6.26	31.13	60.46
71 1/2"	6.30	31.34	60.87
72"	6.34	31.55	61.28
72 1/2"	6.38	31.76	61.69
73"	6.42	31.97	62.10
73 1/2"	6.46	32.18	62.51
74"	6.50	32.39	62.92
74 1/2"	6.54	32.60	63.33
75"	6.58	32.81	63.74
75 1/2"	6.62	33.02	64.15
76"	6.66	33.23	64.56
76 1/2"	6.70	33.44	64.97
77"	6.74	33.65	65.38
77 1/2"	6.78	33.86	65.79
78"	6.82	34.07	66.20
78 1/2"	6.86	34.28	66.61
79"	6.90	34.49	67.02
79 1/2"	6.94	34.70	67.43
80"	6.98	34.91	67.84
80 1/2"	7.02	35.12	68.25
81"	7.06	35.33	68.66
81 1/2"	7.10	35.54	69.07
82"	7.14	35.75	69.48
82 1/2"	7.18	35.96	69.89
83"	7.22	36.17	70.30
83 1/2"	7.26	36.38	70.71
84"	7.30	36.59	71.12
84 1/2"	7.34	36.80	71.53
85"	7.38	37.01	71.94
85 1/2"	7.42	37.22	72.35
86"	7.46	37.43	72.76
86 1/2"	7.50	37.64	73.17
87"	7.54	37.85	73.58
87 1/2"	7.58	38.06	73.99
88"	7.62	38.27	74.40
88 1/2"	7.66	38.48	74.81
89"	7.70	38.69	75.22
89 1/2"	7.74	38.90	75.63
90"	7.78	39.11	76.04
90 1/2"	7.82	39.32	76.45
91"	7.86	39.53	76.86
91 1/2"	7.90	39.74	77.27
92"	7.94	39.95	77.68
92 1/2"	7.98	40.16	78.09
93"	8.02	40.37	78.50
93 1/2"	8.06	40.58	78.91
94"	8.10	40.79	79.32
94 1/2"	8.14	41.00	79.73
95"	8.18	41.21	80.14
95 1/2"	8.22	41.42	80.55
96"	8.26	41.63	80.96
96 1/2"	8.30	41.84	81.37
97"	8.34	42.05	81.78
97 1/2"	8.38	42.26	82.19
98"	8.42	42.47	82.60
98 1/2"	8.46	42.68	83.01
99"	8.50	42.89	83.42
99 1/2"	8.54	43.10	83.83
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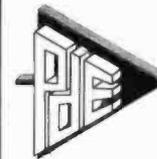
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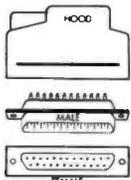


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7414	.29	7495	.79
7415	.39	7496	.79
7416	.39	7497	.399
7417	.39	7498	1.19
7418	.19	7499	.39
7419	.49	7500	.49
7420	.39	7501	.49
7421	.39	7502	.49
7422	.49	7503	.49
7423	.39	7504	.49
7424	.39	7505	.49
7425	.39	7506	.49
7426	.39	7507	.49
7427	.39	7508	.49
7428	.49	7509	.49
7429	.39	7510	.49
7430	.39	7511	.49
7431	.39	7512	.49
7432	.39	7513	.49
7433	.39	7514	.49
7434	.39	7515	.49
7435	.39	7516	.49
7436	.39	7517	.49
7437	.39	7518	.49
7438	.39	7519	.49
7439	.39	7520	.49
7440	.39	7521	.49
7441	.39	7522	.49
7442	.39	7523	.49
7443	.39	7524	.49
7444	.39	7525	.49
7445	.39	7526	.49
7446	.39	7527	.49
7447	.39	7528	.49
7448	.39	7529	.49
7449	.39	7530	.49
7450	.39	7531	.49
7451	.39	7532	.49
7452	.39	7533	.49
7453	.39	7534	.49
7454	.39	7535	.49
7455	.39	7536	.49
7456	.39	7537	.49
7457	.39	7538	.49
7458	.39	7539	.49
7459	.39	7540	.49
7460	.39	7541	.49
7461	.39	7542	.49
7462	.39	7543	.49
7463	.39	7544	.49
7464	.39	7545	.49
7465	.39	7546	.49
7466	.39	7547	.49
7467	.39	7548	.49
7468	.39	7549	.49
7469	.39	7550	.49
7470	.39	7551	.49
7471	.39	7552	.49
7472	.39	7553	.49
7473	.39	7554	.49
7474	.39	7555	.49
7475	.39	7556	.49

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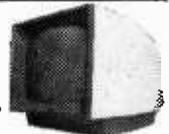
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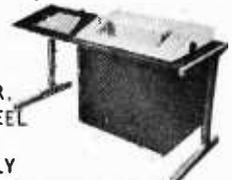
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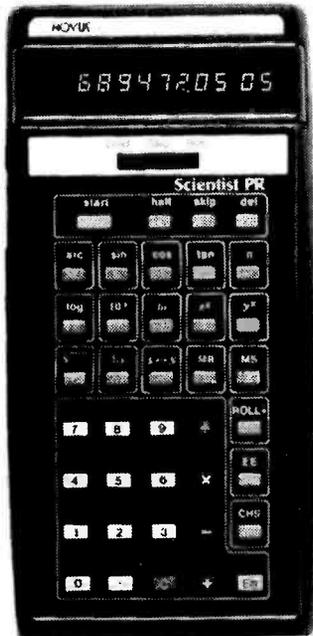
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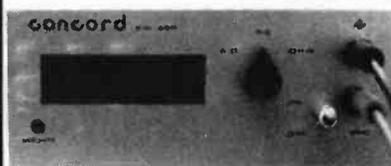
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7427	.25	74193	.70 *	74LS10	.29	4049	.35 *
7437	.20	74194	.85	74LS27	.30	4050	.35 *
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7440	.12 *	74198	1.25	74LS75	.65	4068	.35
7441	.65 *	9602	.50 *	74LS151	1.10	4071	.16
7445	.60	9300	.75	74LS153	1.10	4073	.16
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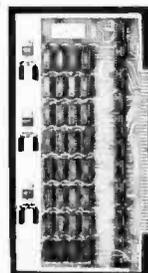
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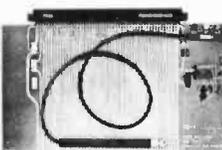
Morrow Sigma100 Front Panel



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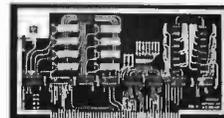
The SIGMA-100 Minicomputer CPU Board (with integral front panel) gives you this control...run your program, or step it at any rate from 1 to 1000 steps per minute. You can stop the machine to examine and alter processor registers, memory locations, and I/O devices---there is special firmware to keep the CPU from going to sleep. You can also monitor all of the above during execution of a program as well. Everything is front panel controlled by your fingertips through a 12 pad keyboard; octal data reads out on 7 segment readouts. DOCUMENTATION: \$5. **\$250**

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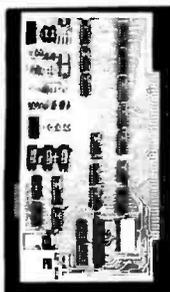


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2704	512 x 8 Bit	450 ns TS Erasable	30.00
2708	1024 x 8 Bit	450 ns TS Erasable	35.00
5201AQ	256 x 8 Bit	1 us TS Erasable	9.00
5201AQ	512 x 8 Bit	1 us TS Erasable	12.00
1067611-5	256 x 4 Bit	60 ns TS	3.25
748387	256 x 4 Bit	65 ns TS	9.00
748387	256 x 4 Bit	65 ns OC	9.00
82838R	32 x 8 Bit	50 ns OC	4.00
828129R	256 x 4 Bit	60 ns TS	4.25
82738	32 x 8 Bit	50 ns OC	4.50

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8038	VCO		4.50
MC4024	Dual VCO		2.75
580	VCO-Function		2.00

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2513	5x7 5 Line	CM2140 Upper case	6.75
MC6571	7x9 7 Line		12.00
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SHIFT REGISTERS			
DYNAMIC			
1401AN	1024 x 1 Bit	2.5 MHz	3.00
2505K	512 x 1 Bit	2.5 MHz	3.00
STATIC			
MM506	100 x 2 Bit		.80
2510A	50 x 2 Bit	1.5 MHz	1.00
2510B	32 x 2 Bit	2.0 MHz	1.00
2533V	1024 x 1 Bit	1.5 MHz	2.00
TM53002	50 x 2 Bit	1.0 MHz	1.00
TM53112	32 x 2 Bit	2.0 MHz	3.95
MM505A	1024 x 1 Bit	1.5 MHz (18 pin)	2.00

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MM0262N	5 Mic. High MOS Clock Driver		3.00
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NB797	Tri State Hex Buffer		1.15
DBR97	Tri State Hex Inverter		1.00
1488	KS232 Quad Line Driver		1.05
1489	KS232 Quad Line Receiver		1.05
D-5207A	Quad NAND to MOS Driver		2.50
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2107B	1024 x 1 Bit	200 ns	6.50
2107C	4 1024 x 1 Bit	270 ns	5.00
2107D	4 1024 x 1 Bit	350 ns	4.50
4008P-6	1024 x 1 Bit	500 ns	2.75
4050AL	4096 x 1 Bit	300 ns	1.00
MM5261	1024 x 1 Bit	400 ns	5.00
MM5262	2048 x 1 Bit	365 ns	3.00
MM5270	1024 x 1 Bit	200 ns	(18 pin) 1.00
MM5280	4096 x 1 Bit	280 ns	(22 pin) 6.00

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5110A	16 x 1 Bit	110 ns	OC	2.00
1101A	256 x 1 Bit	1 us	TS	1.00
2101	256 x 1 Bit	200 ns	1S	3.00
2102	1024 x 1 Bit	1 us	TS	1.50
2102-1	1024 x 1 Bit	500 ns	1S	1.75
2111A-4	256 x 1 Bit	450 ns	1S (18 pin)	1.15
2112A-4	256 x 1 Bit	450 ns	1S (16 pin)	1.45
2504R	256 x 1 Bit	1 us	TS	1.00
3107	256 x 1 Bit	80 ns	OC	2.95
7459	16 x 1 Bit	240 ns	1S	3.00
748201	256 x 1 Bit	450 ns	1S	4.75
81182A	1024 x 1 Bit	500 ns	1S	2.00
7489	16 x 1 Bit	60 ns	OC	2.25
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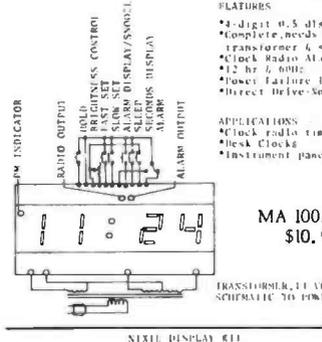
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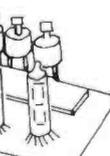


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74LS05	.40
74LS08	.40
74LS10	.40
74LS11	.40
74LS12	.50
74LS120	.40
74LS121	.40
74LS130	.40
74LS132	.40
74LS137	.50
74LS142	.90
74LS151	.60
74LS152	.60
74LS153	.60
74LS154	.60
74LS161	.60
74LS162	.60
74LS163	.60
74LS164	.60
74LS165	.60
74LS168	1.50
74LS169	2.00
74LS174	1.25
74LS175	1.30
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22	Wire Wrap Tin	.95
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24	Wire Wrap Tin	.50
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40	Lo Pro (Open Frame) Tin	.25
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SN7400N	16	SN7459A	25	SN74154N	100
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SN7402N	21	SN7470N	42	SN74156N	99
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SN7407N	29	SN7476N	32	SN74163N	99
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SN7409N	25	SN7479N	41	SN74165A	1.10
SN7410N	18	SN7482N	9.00	SN74166N	1.25
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SN7423N	37	SN7494N	.79	SN74181N	2.49
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CD4017	1.35	74C10N	2.15
CD4019	55	74C10N	2.15
CD4020	1.49	74C10N	3.25
CD4022	1.25	74C10N	3.25
CD4023	25	74C10N	3.00
CD4024	1.50	74C10N	3.25
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74LS03	29	74LS154	1.95
74LS04	29	74LS155	1.95
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74LS06	29	74LS157	2.25
74LS07	29	74LS158	1.95
74LS08	29	74LS159	1.95
74LS09	29	74LS160	2.85
74LS10	29	74LS161	1.95
74LS11	29	74LS162	2.85
74LS12	29	74LS163	2.85
74LS13	29	74LS164	2.85
74LS14	29	74LS165	2.85
74LS15	29	74LS166	2.85
74LS16	29	74LS167	2.85
74LS17	29	74LS168	2.85
74LS18	29	74LS169	2.85
74LS19	29	74LS170	2.85
74LS20	29	74LS171	2.85
74LS21	29	74LS172	2.85
74LS22	29	74LS173	2.85
74LS23	29	74LS174	2.85
74LS24	29	74LS175	2.85
74LS25	29	74LS176	2.85
74LS26	29	74LS177	2.85
74LS27	29	74LS178	2.85
74LS28	29	74LS179	2.85
74LS29	29	74LS180	2.85
74LS30	29	74LS181	2.85
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74LS32	29	74LS183	2.85
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XC209	Orange	4x51	XC256	Orange	4x51	XC556	Orange	4x51		
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14 pin	2.0	19	1.8	28 pin	4.5	44	4.3
16 pin	2.2	21	2.0	36 pin	6.0	59	5.8
18 pin	2.9	25	2.7	40 pin	6.3	62	6.1
22 pin	3.7	38	3.5				
14 pin	5.27	25	24	28 pin	9.99	90	8.1
16 pin	3.0	27	25	36 pin	1.39	1.26	1.15
18 pin	3.5	32	30	40 pin	1.59	1.45	1.30
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8 pin	5.30	27	24	24 pin	5.70	63	5.7
14 pin	9.5	32	29	28 pin	1.10	1.00	90
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18 pin	5.2	37	34	40 pin	1.59	1.45	1.30

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8 pin	5.45	31	37	24 pin	5.10	95	85
14 pin	3.9	38	37	28 pin	1.40	1.25	1.10
16 pin	4.3	42	41	36 pin	1.59	1.45	1.30
18 pin	7.5	68	62	40 pin	1.75	1.55	1.40

WIRE WRAP SOCKETS (GOLDO) LEVEL #3

14 pin	3.9	38	37	28 pin	1.40	1.25	1.10
16 pin	4.3	42	41	36 pin	1.59	1.45	1.30
18 pin	7.5	68	62	40 pin	1.75	1.55	1.40

Plastic Push Button Switch

18 AWG Solid Wire - 5' Long

50 (width) X 6.0 (height) X 27 Thread

8 AMP @ 14 Volt - 1 AMP @ 10 Volt

J-188-1 Push On-Push Off .59 .49

J-188-2 Normally Open .59 .49

J-188-3 Normally Closed .59 .49

MINATURE TOGGLE SWITCH

JMT-221 DPDT on/off/on \$1.95

JMT-223 DPDT on/none/on \$1.75

JMT-121 SPDT on/off/on \$1.50

JMT-123 SPDT on/none/on \$1.25

CLIPLITE 8/S1.49

LED MOUNTING SYSTEM use with XC556 LEDs

Specify Colors - Red - Green - Amber - Yellow

DIP SWITCHES

SPST Slide Action

#206-4 (8 pin dip) 4 switch unit \$1.75 ea.

#206-7 (14 pin dip) 7 switch unit \$1.95 ea.

#206-8 (16 pin dip) 8 switch unit \$2.25 ea.

50 PCS. RESISTOR ASSORTMENTS \$1.75 PER ASS'T.

ASST. 1	5 ea.	10 OHM	12 OHM	15 OHM	18 OHM	22 OHM	1/4 WATT 5% - 50 PCS.
ASST. 2	5 ea.	180 OHM	82 OHM	100 OHM	120 OHM	150 OHM	1/4 WATT 5% - 50 PCS.
ASST. 3	5 ea.	680 OHM	220 OHM	270 OHM	330 OHM	390 OHM	1/4 WATT 5% - 50 PCS.
ASST. 4	5 ea.	470 OHM	560 OHM	680 OHM	820 OHM	1K	1/4 WATT 5% - 50 PCS.
ASST. 5	5 ea.	1.2K	1.5K	1.8K	2.2K	2.7K	1/4 WATT 5% - 50 PCS.
ASST. 6	5 ea.	3.3K	3.9K	4.7K	5.6K	6.8K	1/4 WATT 5% - 50 PCS.
ASST. 7	5 ea.	8.2K	10K	12K	15K	18K	1/4 WATT 5% - 50 PCS.
ASST. 8	5 ea.	22K	27K	33K	39K	47K	1/4 WATT 5% - 50 PCS.
ASST. 9	5 ea.	56K	68K	82K	100K	120K	1/4 WATT 5% - 50 PCS.
ASST. 10	5 ea.	150K	180K	220K	270K	330K	1/4 WATT 5% - 50 PCS.
ASST. 11	5 ea.	390K	470K	560K	680K	820K	1/4 WATT 5% - 50 PCS.
ASST. 12	5 ea.	1M	1.2M	1.5M	1.8M	2.2M	1/4 WATT 5% - 50 PCS.
ASST.							

computer display terminal

This display terminal has an integral controller, B/W cathode ray tube and keyboard. The system has a serial I/O interface for communication and an I/O interface for a printer

DISPLAY (P/N 4802-1095-501) FEATURES:

- 17" B/W CRT
- 41 lines of data
- 52 characters per line
- Characters are generated by a diode matrix "graphic" technique
- 21 special push-buttons wired for a program call up
- Brightness Control
- Self-contained power supply

KEYBOARD (P/N 4802-1115-501) FEATURES:

- Reed switch technology
- 54 data keys
- 28 special keys detachable with cable

LOGIC UNIT (P/N 4802-1157-502) FEATURES:

- 1024 by 6 bit core memory
- Printer I/O interface
- Communication I/O interface

POWER: 115V, 50/60 Hz, 500 Watts

WEIGHT: 210 lbs. (including logic unit, keyboard, display and cables.)

FOB LYNN MASS (you pay shipping)
Check with order please.



\$180.00

External logic & power pack not shown.

"AS IS"

4 way cursor control, graphics display.

The story: These are unused terminals made for airport ticketing & seat assignment. After several years of storage they require tinkering to make operable. We have some hints printed such as cleaning PC fingers. One of our customers has this tied into his KIM-1, another has his running with his IMSAI. We have data on this. Should be useable on most common computers. A hell of a deal and all for a paltry \$180.00. Don't be left out as many were on our past VIATRON deal. Sold "as is" all sales final.

WITH COMPLETE DOCUMENTATION

SPECTRA FLAT TWIST

50 conductor, 28 gauge, 7 strands/conductor made by Spectra. Two conductors are paired & twisted and the flat ribbon made up of 25 pairs to give total of 50 conductor. May be peeled off in pairs if desired. Made twisted to cut down on "cross talk." Ideal for sandwiching PC boards allowing flexibility and working on both sides of the boards. Cost originally \$13.00/ft

SP-324-A \$1.00/ft. 10 ft/\$9.00

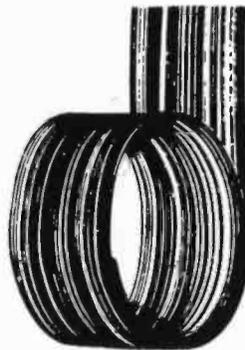
SP-234-A \$1.00 ft 50 cond. 10 ft/\$9.00

SP-234-B .90 ft 32 cond. 10 ft/\$8.00

In tall TO-5 can
DPDT, 24 volts. Brand new.
cost \$16.00 each

SP-134 \$3.00 each 2/\$5.00

TINY
SWITCH
TELEDYNE



WIRE WRAP WIRE

TEFZEL blue #30 Reg. price
\$13.28/100 ft. Our price 100 ft \$2.00;
500 ft \$7.50.

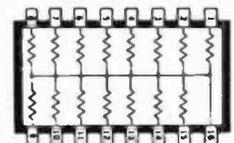
MULTI COLORED SPECTRA WIRE

Footage	10'	50'	100'
8 Cond. #24	\$2.50	9.00	15.00
12 "	22	3.00	11.00 18.00
14 "	22	3.50	13.00 21.00
29 "	22	7.50	28.00 45.00

Great savings as these are about 1/4 book prices. All fresh & new.

Precision 16 pin DIP network as shown.
Each resistor 1K. For pull-up/pull-down
interface networks. Value over \$1.00
each; New, CTS or Beckman

SP-320 pack of 6 \$1.00



Meshna

Please add shipping cost on above. Minimum order \$10
FREE CATALOG SP-9 NOW READY
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KITS BY cybercom

4Kx8 Static Memories

MB-1 MK-8 board, 1 usec 2102 or eq. PC Board \$22 (Not for S-100 Buss)

MB-4 Improved MB-2 designed for 8K "piggy-back" without cutting traces. PC Board\$30
Kit 4K .5 usec...\$129.95 Kit 8K .5 usec...\$199

MB-3 1702A's Eroms, Altair 8800 & Imsai 8080 compatible switched address & wait cycles. 2K may be expanded to 4K. Kit less Proms...\$65 2K Kit...\$105
4K Kit...\$145

MB-6 PC Board\$35

MB-6A 8Kx8 Switched address and wait assignments. Memory protection is switchable for 256, 512, 1K, 2K, 4K and 8K. 91L02A. 5 usec rams. Altair 8800 & Imsai compatible. With battery power option.

Kit...\$250 Assembled & tested\$290

I/O Boards

I/O-2 I/O for 8800, 2 ports, committed pads for 3 more, other pads for EROMS UART, etc.

Kit...\$55 PC Board only\$25

64 x 16 VIDEO BOARD Altair plug compatible display 32 x 16 or 64 x 16 switch selectable. Composite and parallel video ports, upper and lower case with software.

Kit\$179.95

PC Board\$35

SP-1 Synthesizer Board Computer controlled wave forms 9 octaves 1vrms 1/2% distortion includes high level music language. Kit\$250

Altair Compatible mother board. Room for 15 connectors 11" x 11 1/2" (w/o connectors)\$45
With 15 connectors\$105

Altair Extender Board (w/o connectors)\$9.00

With w/w connector\$13.50

100 Pin spec WW or Soldertail both fit Imsai or SSM Mother Board \$5.00 each 10/\$44.

1702A EROM \$ 8.00

programming send hex list 5.00

AY5-1013 UART 6.95

2513 Prime spec. upper or lower case 11.00

8080A prime CPU 25.00

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(AVAILABLE MID-MAY)

MM-1 8Kx8 Ram Board First fully buffered 8K ram board. Addresses and data are fully buffered. Low power rams used. Protect is reset with a single switch. Documentation is available for \$5 which is refunded at time of purchase of board or kit.

MM-1 Kit\$245

Assembled295

PC Board only30

★ 2102AL-4 FULL SPEED MEMORY (500ns) ★

★ Nearly 1/2 less power consumption of even a 21L02A PRIME FROM NEC ★

★ \$2.00 EACH 32 FOR \$1.80 EACH ★
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7405 .24 7476 .30 74166 1.25

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MM5313	4.00	2504V	2.00	2101	4.50	74LS02	.40	74L02	.25	74L86	.75
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MM5554	1.90	2507V	1.25	2602	1.60	74LS04	.45	74L04	.30	74L90	1.50
MM5556	2.50	2509A	2.00	4002-1	7.50	74LS05	.45	74L05	.40	74L91	1.50
MM5055	1.90	2510A	2.00	4002-2	7.50	74LS10	.40	74L06	.30	74L93	1.70
DM8836	.60	2511A	2.80	MM5262	1.00	74LS11	.50	74L08	.40	74L95	1.70
DM8837	1.50	2517V	1.25	7489	2.00	74LS12	.55	74L09	.40	74L98	2.80
80C95	1.10	2518B	1.50	74200	4.95	74LS20	.40	74L10	.30	74L123	1.50
80C97	1.00	2519B	2.80	74S89	3.50	74LS22	.45	74L20	.35	74L154	2.00
80L97	1.50	2521V	1.50	74C89	3.50	74LS27	.45	74L26	.40	74L164	2.50
81L22	1.50	2522V	2.00	74L89	3.50	74LS30	.40	74L30	.40	74L165	2.50
81L23	1.90	2525V	2.80	8T80	2.50	74LS42	1.50	74L32	.45	74L192	1.25
81L51	2.50	2527V	2.80	8T97	2.00	74LS55	.40	74L42	1.50	74L193	1.20
85L52	2.50	2528V	2.80	INTEL		74LS73	.65	74L51	.35	MC4044	2.25
85L63	1.25	2529V	2.80	8216	4.95	74LS74	.65	74L54	.45	N8263	3.50
86L70	1.50	2532B	2.80	8214	8.30	74LS76	.65	74L55	.35	N8826	2.50
86L75	1.90	2533V	2.80	8251	14.50	74LS151	1.55	75L71	.30	DM8131	2.50
86L99	3.50	91L02APC	2.55	8255	14.50	74LS174	2.20	74L73	.55	8T16	2.00
86L12	.80	32 ea.	2.40	1488	1.50	74LS175	1.95	74L74	.55	8T20	2.00
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For large orders please send money order or cashier's check to avoid delays in waiting fr checks to clear.

Check or money order only. California residents add 6% tax. All orders postpaid in U.S. All devices tested prior to sale. Money back 30 day guarantee. Sorry we cannot accept returned ICs that have been soldered to. \$10 minimum order. Prices subject to change without notice.

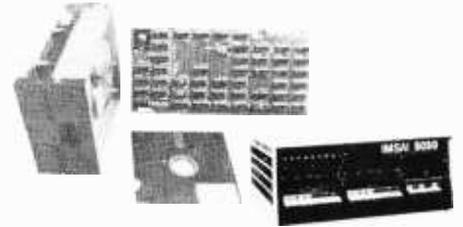


The Proko Paper Tape Reader

A fully TTL buffered optical tape reader for reading 8-level paper tape. It's compatible with nearly all parallel input ports and has both positive and negative strobe outputs. The power (+5V @ 20ma) is derived from your I/O board...just supply a light source (study lamp), grab the tape and pull!

PTR-II Kit \$54.00

Assembled \$68.00



Now, add mass storage to your micro with the North Star floppy disk system. Comes with cables, S-100 controller with bootstrap rom, and powerful extended basic. (Basic supports sequential and random access disk files.)

North Star Disk Kit\$899

IMSAI 8080 Kit with 22 slots\$650

SPECIAL DEAL: IMSAI WITH DISK
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The S.D. Sales Co.
Z-80 CPU Kit
For S-100 Buss
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IMSAI 8080 with 22-SLOT MB
UPGRADED WITH Z-80 CPU \$750⁰⁰

POWER TRANSFORMER
FOR ALTAIR® TYPE BUSS
Designed to deliver +18VDC @ 2A and 5VDC @ 22A with proper bridges and filtering.
29⁵⁰ each

NEW FROM IMSAI
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Full Speed — Low, Low Power
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Please send for xlistor, IC and kit list

Circle 57 on inquiry card.

New Component Values

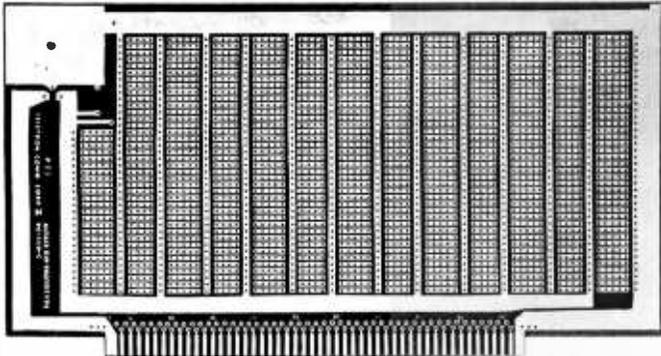
PRIME, NEW 2708 1K x 8 EPROM — **\$ 47.95**

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PROTOTYPE BOARDS



Prototype boards for the S-100 bus are available from many others—but only MINI MICRO MART supplies four different types. Two are wire-wrap versions and two are general-purpose DIP, for either ww or point-to-point wiring. All boards come with a 5V regulator and a heat sink. The two "bus" versions are unique and have circuitry etched on for buffering and address decoding, and include the decoders and necessary tri-state buffers. (Illustrated above is the general-purpose DIP version, MODEL 01-2115.)

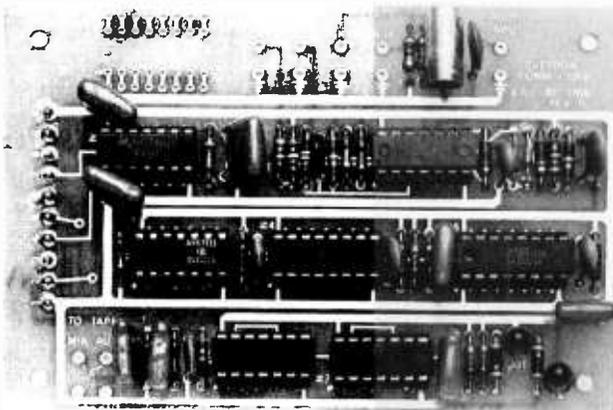
01-2115 General-purpose DIP Prototype Board **\$ 18.95**

01-2116 Wire-wrap Prototype Board **\$ 19.95**

01-2136 Gen.-purpose DIP Bus Interface Bd. . . **\$ 29.95**

01-2112 Wire-wrap Bus Interface Board **\$ 30.95**

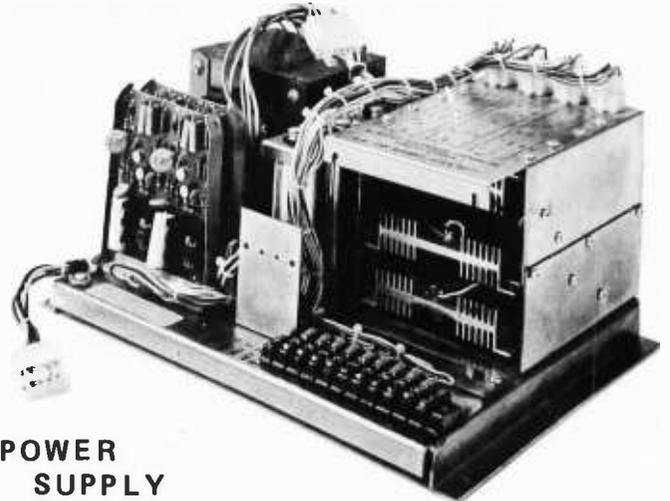
AUDIO CASSETTE INTERFACE



This simple board can be used with any minicomputer or TV terminal that uses a UART with a 16X baud-rate clock. Designed for the St. Louis BYTE standard, it also provides for a tone/no-tone and a HITS interface. Available in kit form, order as —

51-2166 Audio Cassette Interface **\$ 24.95**

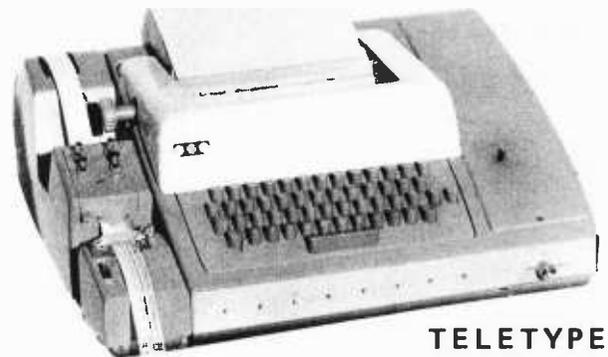
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POWER SUPPLY

Surplus power supply (made by Milwaukee Electronics) removed from used Mohawk Data equipment, excellent condition.... 'Sorry we can't give full details at press time, but we're sure that it will supply all the +5 you'll ever need as well as other voltages. 'Has all the good things — such as over-voltage protection. We hope to have schematics to ship with these units. Limited supply. A REAL VALUE! Order as —

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THE HOTTEST SELLING KIT WE EVER PRODUCED!
You requested it! Our first D.C. operated clock kit. Professionally engineered from scratch. Not a makeshift kluge as sold by others.

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- A. Bowmar Jumbo -.5 inch LED array.
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**50,000 SATISFIED CLOCK
KIT CUSTOMERS CANNOT
BE WRONG!**

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Z-80 CPU 49.95
82S129 1K PROM 2.50

1702A 2K EPROM
We tell it like it is. We could have said these were factory new, but here is the straight scoop. We bought a load of new computer gear that contained a quantity of 1702 A's in sockets. We carefully removed the parts, verified their quality, and are offering them on one heck of a deal. First come, first served. Satisfaction guaranteed! U.V. Erasable. **NEW PRICE! \$2.95 ea.**
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**21L02-1 1K LOW POWER 500 NS
STATIC RAM** Time is of the essence!
And so is power. Not only are our RAM's faster than a speeding bullet but they are now very low power. We are pleased to offer prime new 21L02-1 low power and super fast RAM's. Allows you to **STRETCH** your power supply farther and at the same time keep the wait light off.
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S.D. SALES EXCLUSIVE!
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- KIT FEATURES:**
- A. 60HZ output with accuracy comparable to a digital watch.
 - B. Directly interfaces with all MOS clock chips.
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\$12.95 MOS 6 DIGIT UP-DOWN COUNTER \$12.95
40 PIN DIP. Everything you ever wanted in a counter chip. Features: Direct LED segment drive, single power supply (12 VDC TYPE.), six decades up/down, pre-loadable counter, separate pre-loadable compare register with compare output, BCD and seven segment outputs, internal scan oscillator, CMOS compatible, leading zero blanking, 1MHZ. count input frequency. Very limited quantity! WITH DATA SHEET

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74LS00-49c	7413-50c	7453-19c	74LS90-95c	74154-1.00
7402-19c	7416-69c	7473-39c	7492-75c	74157-75c
74LS02-49c	7420-19c	7474-35c	7493-69c	74161-95c
7404-19c	7430-19c	74LS74-59c	7495-75c	74164-1.10
74LS04-29c	7432-34c	7475-69c	7496-89c	74165-1.10
74S04-44c	7437-39c	7476-35c	74121-38c	74174-95c
74LS04-49c	7438-39c	7480-49c	74123-65c	74181-2.50
7406-29c	7440-19c	7483-95c	74132-1.70	74191-1.25
7408-19c	7447-85c	7485-95c	74S138-1.95	74192-1.25
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TTL INTEGRATED CIRCUITS				74195-69c

1000 MFD Filter Caps
Rated 35 WVDC Upright style with PC leads. Most popular value for hobbyists. Compare at up to \$1.19 ea. from franchise type electronic parts stores. S.D. Special 4/\$1.

Slide Switch Assortment
Our best seller. Includes miniature and standard sizes; single and multi-position units. All new, first name brand. Try one package and you'll reorder more! Special 12/\$1.00

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1/4W 5% & 10% PC leads. A good mix of values. 200/\$2.

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PC leads. At least 10 different values. Includes .001, .01, .05, plus other standard values. 60/\$1.00

\$9.95 KIT

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AC XFMR - 1.50
Do not confuse with Non-Alarm kits sold by our competition!
Eliminate the hassle - avoid the 5314!

SIX DIGIT ALARM CLOCK KIT
We made a fantastic kit even better. Redesigned to take advantage of the latest advances in I.C. clock technology. Features: Litronix Dual 1/2" displays, Mostek 50250 super clock chip, single I.C. segment driver, SCR digit drivers. Greatly simplified construction. More reliable and easier to build. Kit includes all necessary parts (except case). P.C.B. or XFMR optional.
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Motorola SCR
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The Bomb of March

The BOMB of March has fallen. In the tabulation of the stacks of cards returned with article evaluations, Jack Breimeir and Ira Rampil came out first, and receive the \$100 prize for their article "The Digital Cassette Subsystem, Part 2." Second prize winner in the tally came up as a tie between Steve Ciarcia's "Try This Computer on for Size" and Thomas R Buschbach's "An Inexpensive Joystick Interface." Steve and Thomas will each receive a \$50 bonus check. ■

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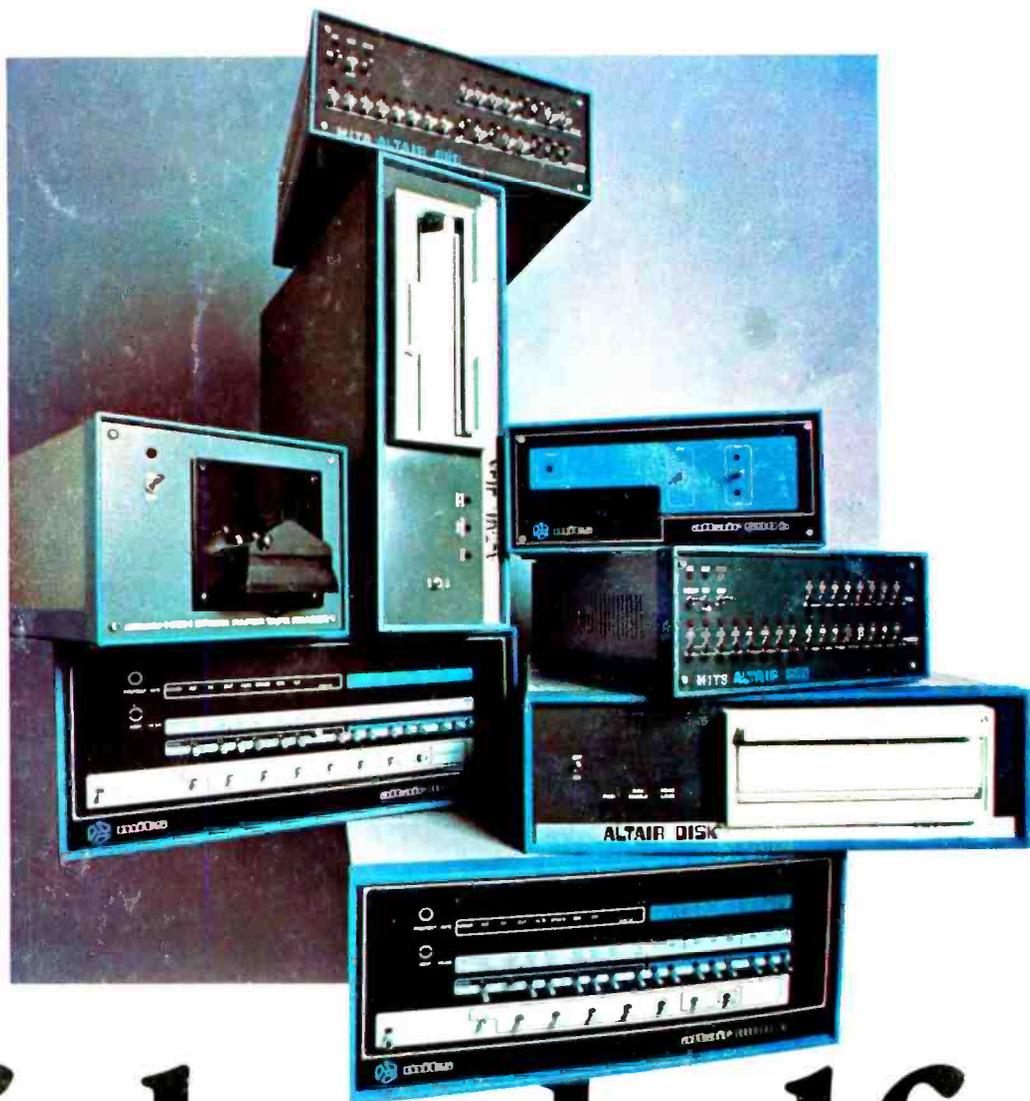
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