

- JANUARY
- * RADIO ENGINEERING
- * EMERGENCY COMMUNICATION SYSTEMS

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- * COUPLED CIRCUIT DESIGN
- * AERONAUTICAL COMMUNICATIONS
- ★ TBA CONFERENCE REPORT
 - * TELEVISION ENGINEERING



AMPEREX

the high performance tube



0

WATER AND AIR COOLED TRANSMITTING AND RECTIFYING TUBES



With the ILLINOIS TOOL WORKS, as with many other leading concerns working with electronic tubes, it's the "Amperextra" of longer life and low-cost efficiency that has made our products a first and exclusive choice. AMPEREX

pioneered in the field of tubes for industrial applications. We are familiar with the needs of industry, and we have the tubes to meet all requirements. Consult AMPEREX for assistance with your present or postwar problems.

> **IMPORTANT!** AMPEREX tubes are now available through leading radio equipment distributors. This new arrangement may save valuable time for busy engineers by enabling them to obtain many of our standard tube types from their local supply sources.

AMPEREX ELECTRONIC CORPORATION

79 WASHINGTON STREET . . . BROOKLYN 1, N. Y. Export Division: 13 E. 40th St., New York 16, N. Y., Cables: "Arlab"

THE WAR ISN'T OVER YET ... BUY AND HOLD MORE WAR BONDS

108 SERIES Amplifiers

WITH MOUNTING ACCESSORIES

TYPE 108-B two-stage Amplifier provides transformer input impedances for either 30 or 250 ohms with nominal output impedance 500 or 8 ohms. Variable gain 65/105 db. with electronic volume control. Frequency response better than ±1 db. 30/16,000 c.p.s. Power output +43 V.U. (20 watts) with less than 5% RMS harmonic content. Noise level full gain 56 db. below full output.

THE 108 SERIES consist of four different amplifiers available simply by changing one or two small input panels on the master chassis. Except for these input panels all amplifiers have the same transmission characteristics. Input impedance, gain and noise level depending on types listed below.

These units are designed for the highest type audio service having gain-frequency characteristics better than ± 1 db. 30/16,000 c.p.s. Power output +43 V.U. (20 watts) with less than 5% RMS harmonic content.

TYPE 108-A two-stage Amplifier provides transformer input for either 600 ohm or bridging. 600 ohm input fixed gain 61 db. Bridging input variable gain 6/46 db. Noise level 68 db. below full output.

TYPE 108-B as illustrated and described above.

TYPE 108-C combines the input channels of the 108-A and 108-B Amplifiers. Channel 1—600 ohm input variable gain 20/60 db. Bridging input variable gain 2/42 db. Channel 2—high gain 30/250 ohm input variable gain 62/102 db. with electronic volume control. Noise level 56 db. below full output.

TYPE 108-D two-channel each 30/250 ohm input. Either channel variable gain 62/102 db. with electronic volume control. Noise level 56 db. below full output.

MOUNTING ACCESSORIES

TYPE 202-A Wall Mounting Cabinet permits universal installation of 108 Series Amplifiers to any flat surface. Well ventilated and designed for maximum accessibility, servicing and convenience of installation. Standard aluminum gray finish.

TYPE 9-A Modification Group permits 108 Series Amplifiers to mount on standard 19" telephone relay racks. Occupies 7" rack space. Allows servicing from front of rack. Standard aluminum gray finish.



NEW YORK 37 W. 65 St., 23 SAN FRANCISCO 1050 Howard St., 3 www.americanradiohistory.com LOS ANGELES 1000 N. Seward St., 38

We See.

AN EPOCHAL EVENT LAUNCHED the New Year . . . the FCC reallocation proposal covering the 25 to 30,000-mc spectrum . . . a proposal that provides for the greatest expansion in communications history. The FCC plan calls for new frequency assignments to railroads, rural telephones, general mobile services (trucks, buses, taxicabs, ambulances) and walkie-talkies for Mr. and Mrs. Public.

F-m received 90 channels between 84 and 102 mc. Twenty of these channels (84 to 88 mc) are for educational stations. A band of from 102 to 108 mc was set aside for f-m expansion. Television was given six 6-mc channels from 44 to 84 mc and six 6-mc channels from 180 to 216. And to permit development of high definition and color pictures, bands between 480 and 960 mc were also set aside for experimental television. The air services received substantial grants . . . almost exactly as requested. In offering these frequencies, FCC officials said . . . "aviation operations are wholly dependent upon the use of radio for insuring the safety of life and property in the air.'

Twenty-two channels between 25 and 28 mc were proposed for relay broadcast operations. Studio-transmitter links were moved to the 900 to 960 mc band. Here incidentally f-m will be used.

The 460- to 470-mc band was set aside for a walkie-talkie service which will be available to practically anyone. Operators will have to abide by only the minimum requirements of the Communications Act and a few traffic rules. But no technical knowledge will be necessary, says the FCC. Doctors, farmers, sportsmen, department stores, boatmen and countless others will find this quite a handy service.

Police, fire, forestry and railroad services were assigned the 148- to 162-mc bands.

The proposal in its present or modified form will become a final program soon after oral hearings with industry specialists are held during the middle of February.

Congratulations to the FCC for a comprehensive appraisal and sturdy approach to a complex problem.





JANUARY, 1945

VOLUME 25 MRFR

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In the control room of AES at Noumea, New Caledonia. (Courtesy U. S. Army Signal Corps)

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JANUARY

Published in the Interests of Better Sight and Sound

1945

Electronic Devices Broaden Sylvania's Service to Industry

The manufacture of electronic equipment for certain specialized communications and industrial applications is an important phase of Sylvania service. Manufacture of this type of equipment is carried



An electronic device undergoes test in the laboratories of Sylvania's Industrial Apparatus Plant.

on in a separate Industrial Apparatus Plant located at Williamsport, Pa.

This aspect of Sylvania's activities is a natural outgrowth of the company's intensive experience in the design and application of electron tubes.



That Sylvania Tungsten Lamps are standard equipment for signaling purposes on many telephone switchboards? They are made in a range of electrical characteristics for use in any type of switchboard.

That Sylvania Near Ultra-Violet Lamps activate the fluorescent dials on airplane instrument panels? Lamps are small, compact, designed to operate from a 24-28 volt direct current source.

*

Sylvania Begins Survey of Public Interest in Television Receivers Findings Will Assist Manufacturers in Gaging Markets, Determining Price Range

Thousands of personal interviews and an intensive advertising campaign in the pages of leading consumer publications form the twin phases of a comprehensive survey which Sylvania is launching to gage the interest of consumers in the purchase of television sets, and to learn the extent of the

LOCK-IN TUBES IDEAL FOR UHF

The trend toward the use of ultra-high frequencies brings to the fore the outstanding advantages of Sylvania's Lock-In Tubes. While the name of this line of tubes has tended to emphasize the physical details of mounting, one of the chief motivating forces in their design was the desire of Sylvania engineers to improve the electrical characteristics of tubes, particularly at the higher frequencies.

The Lock-In feature itself has been responsible for the extensive use of these tubes, particularly in automobile radios; electrical features point to wide utilization in television and FM.



"I wonder if I could have your views on what the postwar radio will be like."

potential market for receivers in various selling price ranges. The results of this survey are expected to be of great value in guiding the planning of the manufacturers of television sets.

Television, moreover, is but one of the aspects which will be covered in this



The type of set people prefer-floor or table model, radio only or radiophonograph combination — will also be studied in the Sylvania survey.

nation-wide poll. Consumers will also be queried on such points as their interest in FM; the desirability of short-wave bands; reaction to push button tuning. The reasons why people decide on new set purchases will also come in for scrutiny.

As the survey progresses, findings will be reported from time to time in future issues of SYLVANIA NEWS.

SYLVANIA SELECTRIC

SYLVANIA ELECTRIC PRODUCTS INC., Radio Division, Emporium, Pa. MAKERS OF RADIO TUBES, CATHODE RAY TUBES, ELECTRONIC DEVICES, FLUORESCENT LAMPS, FIXTURES, ACCESSORIES, INCANDESCENT LAMPS COMMUNICATIONS FOR JANUARY 1945 • 3

Now, EXTRA HU MIIIIY PROTECTION IS STA

designed for tropical conditions unbeatable on ANY job

Standard Sprague Koolohm Wire Wound Resistors now offer the same high degree of humidity protection formerly obtainable only on special order to match exacting military specifications. This construction, newly adopted as standard, includes a glazed ceramic outer shell and a new type of end seal. These features give maximum protection against even the most severe tropical humidity conditions. Type numbers remain the same except for the fact that the letter "T' has been added to designate the new standard construction.

WOUND WITH CERAMIC

INSULATED

WIRE

Thus, again, Sprague leads the way in practical, truly modern wire wound resistor construction. Your job of resistor selection is greatly simplified. No need to study and choose between types or coatings. One type of Koolohms, the standard type, does the job -under any climatic condition, anywhere in the world!

SPRAGUE ELECTRIC COMPANY, North Adams, Mass. (formerly Sprague Specialties Co.)



even years ago we started out as a west coast, distributor of aircraft parts. We named our company Aircraft Accessories Corporation. Then we started developing aircraft hydraulics. The next thing we knew we were full-fledged manufacturers-and out of the parts business. Later, someone came along with an embryo electronic plant in Kansas City. Being young and ambitious-we bought it. To everyone's amazementand somewhat to our own—we made it grow. And pay.

NNAM NEW HORIZONS

We're still young, ambitious. Our explorations in hydraulics, electronics and other fields promise post-war growing pains. And so we've outgrown our name. Aircraft Accessories Corporation no longer adequately describes our operations. We couldn't think of a name that did. So we coined one: Aireon. It's a name that's partly aircraft, partly electronics; but it will be largely what we make it. We hope - and intend - to make Aireon worthy of a place among America's most honored corporate and trade names.



CHICAGO

MANUFACTURING CORPORATION Formerly AIRCRAFT ACCESSORIES CORPORATION

Radio and Electronics • Engineered Power Controls

KANSAS CITY .

BURBANK

a New and Superior DIAL LIGHT SOCKET

Tensile strength of leads and connections far in excess of requirements.

Tough, plastic shell molded around bracket providing a secure bond with mechanical strength far beyond any normal requirement.

Rounded edge will not cut or fray wire insulation.

Voltage Breakdown between contacts-1200 Volts. Voltage Breakdown to ground --5000 Volts.

Lug on contact fits in groove in shell so that contact cannot be turned or twisted when inserting lamp.

Center contact mounted so that it cannot protrude from shell and short on chassis when lamp is removed.

Plastic shell is recessed for contacts, which cannot be pushed or pulled out of position.

Stronger, tougher, neavy walled plastic shell,-

A variety of different mounting bracket styles available, suitable for practically any mounting.

ELECTRIC

CO.

CHICAGO 47, ILLINOIS

For Your Present and Post-War Production

Lenz Dial Light Sockets have always been known for their superior mechanical , qualities and electrical characteristics.

Now these sockets are still further improved, with even greater mechanical strength. A stronger, tougher plastic shell is attached to the bracket with a new type of construction that provides a virtually unbreakable bond between shell and bracket. Its excellent electrical characteristics are maintained. Consider these Lenz Dial Sockets for your present and post war production. Write for sample today.

MANUFACTURING

LENZ

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40TH ANNIVERSARY

1904-1944

This year Lenz celebrates its 40th year of service to the communications industry.

4 TIMES

ACTUAL

SIZE

In **Business**

Since 1904

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•

A NEW -hp- AUDIO OSCILLATOR

3 OUTSTANDING NEW FEATURES: 1. ACCURATE CONTROL DIAL



New friction driven dial provides hair-line accuracy



Rugged chassis construction on heavy cast frame

EASY TO READ • NO PARALLAX 2. RUGGED CONSTRUCTION 3. IMPROVED CIRCUIT



The Model 200-I is a resistance-tuned audio oscillator designed to provide high stability and accuracy for use in frequency measurements. It has a range

from 6 cps to 6000 cps,* divided into six frequency ranges as follows: 6 to 20, 20 to 60, 60 to 200, 200 to 600, 600 to 2000 and 2000 to 6000 cps. Each of these ranges has an individual frequency adjustment so that the instrument may be set to a frequency standard such as the *-hp*- Model 100-B.

The large, 6-inch diameter, main frequency dial is calibrated over approximately 300 degrees, making possible a large number of calibrated points to cover the entire range. The dial itself rotates behind a fine wire locator which is visible through an opening in the panel. Parallax is completely eliminated and calibrations are spread over an effective scale length of nearly eight feet. Fast and extremely accurate settings are made easily with this dial. There are two manual controls: one direct action, and the second for vernier adjustments. An electronically regulated power supply is included to assure greatly improved stability.

The Model 200-I -*hp*- Audio Oscillator is but one among many new -*hp*- instruments which are to make public appearance as the cloak of military secrecy is removed. Preliminary technical information is available on this instrument now ...write for it! And watch for the early release of other new -*hp*- instruments.

> *This frequency coverage was selected for interpolation work. Other frequency ranges can be supplied from 6 cps to 100 kc on special order.

HEWLETT - PACKARD COMPANY Box 930E - Station A - Palo Alto, California

OHMITE Many Types and Sizes RESISTORS Assure the Right Resistor for Each Control Job

Many critical control problems are being readily and successfully solved with Ohmite Resistors. That's because the extensive range of Ohmite types and sizes makes possible an almost endless variety of regular or special units to meet each need best.

Ohmite core sizes range from $2^{1}/_{2}$ " diameter by 20" long to $\frac{5}{16}$ " diameter by 1" long. Wide selection of stock units are available.

These rugged resistors have proved their worth under toughest operating conditions, in every field of action. Ohmite engineers are glad to help on today's and tomorrow's control problems.

OHMITE MANUFACTURING COMPANY 4869 Flournoy Street * Chicago 44, U.S.A.

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TYPICAL UNITS

Fixed Lug "Dividohm" Wire Lead "Corrib" Ferrule Edison Base Precision Bracket Non-Inductive Tapped Cartridge Strip

OHMIT

Send for Catalog and Engineering Manual No. 40 Write on company letterbead for this complete, belpful guide on resistors, rheostats, tap switches. Address OHMITE, 4869 Flournoy, Chicago 44.



THEY SAID IT COULDN'T BE DONE !

Back in 1938, Hytron began designing new dies and converting production machinery for the first BANTAM GT tubes. The industry said in effect: "You're crazy; it won't work. You can't telescope standard glass tubes to BANTAM size and get the same results." Beam tetrodes, such as the 50L6GT, particularly were considered impossibilities. The intense heat developed during normal operation would warp the elements and crack the small glass bulb. But Bruce A. Coffin, originator of the BANTAM GT, stuck to his guns. In a few short years, Hytron developed over fifty GT types. The GT became the most popular receiving tube.* Short leads, low capaci-

tances, advantages of shorter bombardment at lower temperatures, ruggedness of compact construction plus both top and bottom mica supports, smaller size, standardized envelopes and bases — all contributed to that popularity.

The BANTAM GT permitted new space economies in pre-war receivers. Only its universal acceptance as standard by all manufacturers makes possible fulfillment of the Services' demands for receiving tubes. In increasing numbers, as this war draws to its ultimate conclusion, Hytron will continue to supply you with the popular BANTAM GT tubes which everyone said just couldn't be made.

*1941 industry production figures: GT - 52,000,000; metal - 27,000,000; standard glass, G, and loctal - 56,000,000.





A Word About Reconversion

R^{ECONVERSION already is getting started in a few plants. For others, it may be just "around the corner", but Radio still is completely absorbed in its wartime job.}

That's the situation at Rola today. The things made here . . . transformers, coils and other intricate parts for Military Communications . . . still are being required in gigantic quantities, and since Rola is one of the few plants of its kind equipped to make those things, the obligation to produce *in maximum amounts* cannot be slighted. our old customers the kind of service they have learned to expect from Rola ... all the experimental models, all the technical assistance and all the other things we used to provide. This we should regret, for we are proud of our quarter-century reputation for Service, but there is no alternative and we hope our friends in the Radio Industry will understand our present position.

No one can predict how long this intervening period may be, but Rola's reconversion ... when it comes ... will be speedy, and at that time set makers again can look to Rola for the "Finest in Sound Reproducing Equipment."

This means we may not be able, now, to give

THE ROLA COMPANY, INC. • 2530 SUPERIOR AVENUE, CLEVELAND 14, OHIO



MAKERS OF THE FINEST IN SOUND REPRODUCING AND ELECTRONIC EQUIPMENT 10 • COMMUNICATIONS FOR JANUARY 1945



The Greeks gave us a word for it ... now we give it to you

WHEN Sperry first developed its velocity-modulated, ultra-highfrequency tube, the word "KLY-STRON" was registered as the name of the new device.

This name — from the Greek, as coined by scientists of Stanford University — is an apt description of the bunching of electrons between spaced grids within the tube.

"Klystron" is a good name. So good, that it has come into widespread use as the handy way to designate *any* tube of its general type, whether a Sperry product or not.

This is perfectly understandable. For the technical description of a Klystron-type tube is unwieldy, whether in written specifications, in conversation, or in instructing members of the Armed Forces in the operation of devices employing such tubes.

These conditions have prompted many requests from standardization agencies—including those of the Army and Navy—for unrestricted use of the name Klystron. In the public interest, Sperry has been glad to comply with these requests ...

From now on, the name KLYSTRON belongs to the public, and may be used by anyone as the designation for velocity-modulated tubes of any manufacture.

Sperry will, of course, continue to make the many types of Klystrons it now produces, and to develop new ones.

On request, information about Klystrons will be sent, subject to military restrictions.

SPERRY GYROSCOPE COMPANY, INC. GREAT NECK, N.Y.

Division of the Sperry Corporation

LOS ANGELES • SAN FRANCISCO • NEW ORLEANS HONOLULU • CLEVELAND • SEATTLE



In every branch of the Army and Navy, Advance Relays are dependable components of many types of electrical equipment. Factory production is still all-out for War, but Engineering Service for Post-War products is available *now*.





CO-OPERATIVE ENGINEERING FOR NEW RELAY APPLICATIONS

SIMPLIFY your product development and manufacturing problems by calling on Advance Engineers for the close co-operation they are prepared to give you *immediately*. This service not only applies to adaption of one or more of the forty basic types of Advance Relays, which are usually directly applicable to most modern electrical control requirements, but may be utilized for the development of special test models. In any event, you will have the best of relays, exactly as you want them. Write today

Advance Relays

ADVANCE ELECTRIC & RELAY CO. 1260 WEST SECOND STREET • LOS ANGELES 26, CALIFORNIA





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Crystal Blanks TO YOUR SPECIFICATIONS

C rystal Products Company can supply Quartz Crystal blanks in any of the three stages of manufacture: (1) "roughsawed" blanks, (2) "semi-finished" blanks, and (3) "electrically finished" blanks.

"Rough-sawed" blanks are cut to the specified angles and roughly sawed to dimensions.

"Semi-finished" blanks are blanks which have been brought to approximate dimensions by machine lapping, allowance being made for final hand finishing.

"Electrically-finished" blanks are finished by hand to the frequency desired and electrically tested.

All crystal blanks are cut to specifications from selected Brazilian quartz and guaranteed free from all impurities such as optical twinning, electrical twinning, bubbles, fractures, scratches, mineral inclusions, and other mechanical and electrical imperfections. Dimensions, temperature coefficients, and frequencies are guaranteed within specifications listed. Send us your holders for replacement of crystal blanks to exact specifications. ODUCT

1519 McGee Street, Kansas City, Mo. Producers of Approved Precision Crystals

for Radio Frequency Control

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A Complete Line of Standard and Hermetically Sealed TRANSFORMERS

Secret

Many products and processes of Hudson American's war time engineering are still strictly secret — and must remain so for the duration. However, there is no secret to Hudson American's success in transformer engineering.

Not Secret

Only the best in men and machines can provide the best in transformer design and construction.

These Hudson American transformers have proven reliable under the tests of battle — the Arctic cold of bomber flights, the brutal beatings of amphibious landings, and the dust and mud of jungle trails.

Hudson American transformers have successfully met these war imposed tests with the same absolute reliability that they show in industrial applications.

HUDSON AMERICAN CORPORATION 25 West 43rd Street • New York 18, N. Y.

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THE NEW

1945 HUDSON

AMERICAN

to the RADIO INDUSTRY

Whether Amplitude Modulation . . . Frequency Modulation . . . or Television — dependability is a *must* for all broadcast equipment.

Federal broadcast equipment has earned a reputation for that dependability because *it stands up*.

For more than thirty-five achievement-studded years ... from the Poulsen Arc to the new CBS Television Station ... Federal has served the broadcast industry with superior equipment.

Federal's background includes such milestones of electronic progress as the 1000 Kw Bordeaux Transmitter; Micro-ray, the forerunner of modern television technique; and the first UHF multi-channel telephone and telegraph circuits, part of a world-wide communications system . . .

All this, plus the war-sharpened techniques that are the result of ability and experience, combine to give you craftsmanship . . . the kind of craftsmanship that builds dependability into all Federal equipment.

In AM ... FM ... TV ...

and Radio

... your prime need in broadcast equipment is dependability – look to Federal for it.



Newark 1, N. J.

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History of Communications. Number Thirteen of a Serv

MILITARY RADIO COMMUNICATIONS



Today the allied military radio equipments represent the "tops" engineering design. Progress from the spark transmitter of World Way to present-day equipment is, indeed, a far cry. Taking up where the left off December 7, 1941, Universal Engineers, with their added experence with precision military equipment, shall produce for the publi electronic devices not of fantastic design — but of proven utility an quality.

After Victory is ours, radio amateurs, affectionately known as "hams will be back after their experience with military radio equipment wit an even greater desire to operate their own "rigs." It will be then the Universal will again have Microphones and recording component available on dealers' shelves.

 FREE-History of Communications Picture Portfolio. Contains over a dozen pictures suit- able for office, den, or hobby room. Write for your "Portfolio" today.



UNIVERSAL MICROPHONE COMPANY INGLEWOOD, CALIFORNIA

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A liberal choice of types to meet most electrical and mechanical requirements... AFROMON

• Along with pioneering the dry electrolytic capacitor for radio, electronic and motor-starting functions, Aerovox has always maintained an outstanding choice of types.

The new Aerovox Capacitor Catalog now off the press lists 17 types of electrolytics—round-can, square-can, cardboard-case, tubulars, plug-ins, twist-prong base, etc. You will usually find a type listed that precisely meets your capacitance, voltage, mounting, terminal and container requirements. But if your requirements happen to be very unusual, this wide variety of designs enables Aerovox to work out a special type to meet those high-priority needs quickly, satisfactorily, economically.

• Write for Literature ...

Write on your business stationery for latest catalog on electrolytics. Submit that capacitance problem for our engineering collaboration, specifications, quotations.



AEROVOX CORPORATION, NEW BEDFORD, MASS., U. S. A. Export: 13 E. 40 St., New York 16, N. Y. • Cable: 'ARLAB' •

INDIVIDUALLY TESTED

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SALES OFFICES IN ALL PRINCIPAL CITIES In Canada: AEROVOX CANADA LTD., HAMILTON, ONT.



Another SOLA CONSTANT VOLTAGE TRANSFORMER that has an important future in YOUR postwar plans

There may be one vulnerable spot in the design of your equipment that this SOLA Constant Voltage Transformer will correct.

Your customers do not have the stable line voltage called for on your label. They will blame your equipment for inefficient operation, not the fluctuating voltages that really cause it.

Build this SOLA Constant Voltage Transformer into your product and you can be certain that the operating voltages will always be within $\pm 1\%$ of rated requirements regardless of line fluctuations as great as 30%.

This SOLA Constant Voltage Transformer is built to fit your equipment. (Note the small, compact dimensions.) Its low cost will fit your production budget. Its automatic operation will eliminate the need for other costly components, and relieve your customers of the responsibility of making manual voltage adjustments.

There's a powerful sales story behind a product equipped with a SOLA Constant Voltage Transformer.

Constant Voltage Transformers

To Manufacturers:

Built-in voltage control guarantees the voltage called for on your label. Consult our engineers on details of design specifications.

Ask for Bulletin ECV-103

Transformers for: Constant Voltage • Cold Cathode Lighting • Mercury Lamps • Series Lighting • Fluorescent Lighting • X-Ray Equipment • Luminous Tube Signs Oil Burner Janition • Radio • Power • Controls • Signal Systems • Door Bells and Chimes • etc. SOLA ELECTRIC CO., 2525 Clybourn Ave., Chicago 14, III.

Sensitive SAP-ACTION

...in a new, simplified design

The new, simplified construction of the Struthers-Dunn Type 79XAX Sensitive Snap-Action Relay makes it particularly suitable for a wide range of applications because of its ease of adjustment. Snapaction contacts eliminate the erratic, undependable action normally encountered in ordinary sensitive relays when a slowly varying coil current tends to balance the armature tension spring, and to hold closed the normally closed contacts,

The armature of the 79XAX almost completes its travel in either direction before the contacts snap into the new position. This feature permits an unusually broad range of use from vacuum tube circuits, to overcurrent protection, pulsing circuits, and jobs where extremely close differential or extreme sensitivity of operation is required.

The standard adjustment using 60 ampere turns in the coil at approximately .02 watts results in contact pressures of 5 grams with contacts rated 5 amperes, 115 volts a-c; or 0.5 amperes, 115 volts d-c, noninductive. Contact ratings up to 10 amperes, 115 volts a-c may be obtained with 100 or more ampere turns and a corresponding increase in power. A sensitivity of 0.005 watts, with 30 ampere turns, is obtainable with reduced contact pressures and ratings, and at an increase in price of the unit.

STRUTHERS - DUNN INCORPORATED 1321 ARCH STREET, PHILADELPHIA 7, PA.

ERS

A TYPICAL CLOSE DIFFERENTIAL APPLICATION

In using the Struthers-Dunn 79XAX Relay, extremely close differential between pick-up and drop-out may be obtained for potential operation as shown above. The resistor is chosen so that, when the armature closes, the coil current is automatically reduced to a value just sufficient to hold it closed. Any further decrease in voltage will

cause the relay to return to its normal de-energized position as shown.

-NE

DISTRICT ENGINEERING OFFICES: ATLANTA + BALTIMORE + BOSTON + BUFFALO + CHICAGO + CINCINNATI + CLEVELAND DALLAS + DENVER + DETROIT + HARTFORD + INDIANAPOLIS + LOS ANGELES + MINNEAPOLIS + MONTREAL NEW YORK + PITTSBURGH + ST. LOUIS + SAN FRANCISCO + SEATTLE + SYRACUSE + TORONTO + WASHINGTON

TRUT



GEARED TO WAR DEMANDS



Johnson production facilities are flexible - -10 or 10,000 - - standard specifications or special - - repeat items or new - - any plating - - any metal or alloy - - any insulation. If it's metal or insulation or a combination of both, try Johnson first.

Ask for Catalog 968 (E)

E.

F.

JUHNS

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SPECIALISTS IN

- Fixed and Variable Condensers
- Porcelain and Steatite Insulators
- Plugs, Jacks, Clips and Connectors
- Fixed and Variable Inductors
- Radio Frequency Chokes
- Flexible and Rigid Insulated Couplings

MINNES

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OTA

- Antenna Systems and Equipment
- Mycalex Machining and Parts
- Special Insulated Assemblies
- Broadcast Station Equipment

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WASECA

a famous name in Radio

Breeze Flexible Conduit for shielding or ducts

Breeze Flexible Tubing, manufactured in many diameters from a wide variety of metals, is used as ventilation and exhaust ducting in industrial, aircraft, marine and automotive applications. The same basic tubing, with the addition of a braided metal covering, becomes light-weight shielding conduit, used extensively for shielding ignition systems and any electrical circuit to insure dependable radio communication.

Manufactured from a continuous strip of metal, Breeze Flexible Tubing and Conduit can be furnished cut to length, with necessary end-fittings for any conduit installation.



Many different types of interlock construction — plain, packed, and soldered—are available to meet varying use requirements. A few of these are illustrated below.



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Elmac A Grid

Solves grid emission problem!

Grid emission, with the resulting instability of operation, parasitic oscillations and prematurely burned out filaments, has become a thing of the past. The new Eimac "X" grid will not emit electrons even when operated at incandescent heat.

The solution to the problem of grid emission lifts a great barrier which has long stood in the path of electron vacuum tube development and the progress of electronics. Eimac Engineers in developing the "X" grid have made a real contribution... and a very important one ... toward the development of new and more efficient vacuum tubes. It is such heads-up Engineering that has made Eimac first choice of leading electronic engineers throughout the world ... and maintained them in that position year after year.



Expost Agents: FRAZAR & HANSEN, 301 Clay St., San Francisco 11, Calif., U.S. A.

The Science Behind the Science of Electronics is the focusing of all branches of science upon the development and improvement of electron vacuum tubes.



PHYSICS... Especially designed electron microscope enables operator actually to view electron emission.



OPTICS. Photomicrographic studies help achieve perfection in processing.



POWDER METALLURGY...Compounding special alloys.



THERMO-DYNAMICS...Vacuum furnaces beat materials to exceedingly high temperatures.



CHEMISTRY... Experimentation with metallic components and preparation of chemical compounds.



ELECTRONICS... Determining facts about and recording data on vacuum tube capabilities.

A HIGH HONOR FOR THE DEJUR INSPECTION PERS

WAR DEPARTMENT ARMY AIR FORCES MATERIEL COMMAND

Subject: Approved inspection rating DeJur-Amsco Corp. Northern Blvd. at 45th Street Long Island City 1, N. Y.

To:

United States Army Air Forces' Approved Quality Control Rating

"As a result of the efficiency of your inspection personnel and procedures in production of equipment for the Army Air Forces, your company has been given an 'approved' quality control rating"-this is a direct quotation from the letter received by the DeJur-Amsco Corporation from the District Supervisor, Eastern Procurement District, AAF Technical Service Command.

This rating is the Army Air Forces' official recognition of the ability of the DeJur inspection personnel to meet all AAF inspection requirements. And it means that in future fabrication of DeJur products, Air Corps inspectors will not function in the same capacity as heretofore but will stand by merely for supervisory purposes. The men and women inspectors of the DeJur-Amsco Corporation are extremely pleased with this new honor, and pledge unrelenting care in the performance of their duties.

BUY MORE WAR BONDS ... HOLD ON TO THEM

NORTHERN BOULEVARD AT 45th STREET

New York



www.americanradiohistory.com

Long Island City I

PRECISION AT

2,000,000

We ARE not going to tell you here why this tube was made, what it does, or what it is for. The important thing about it to you as a user of radio oscillator, amplifier and

rectifier tubes for communications or induction heating is that it represents the toughest assignment ever handed the electronic tube industry, and that of all tube makers only Machlett perfected the techniques that made the tube possible.

ANOTHER

MACHLET

The tube is sealed-off, vacuum-tight, and operates at 2,000,000 volts, direct current. These and other difficult conditions were essential to assure high and constant power, reduction of heat, and precise focusing of the electron beam.

Electrical and mechanical problems presented by the tube were so severe that some scientists doubted they could be solved, but Machlett, drawing upon its long experience, met every requirement in a little over two years.

This is significant to you because every electronic tube, whether it produces X-rays, or radio waves, or is a rectifier, depends for its success in your service upon correct design, proper vacuum, adequate insulation, and precision-made parts, to assure precise control of the electrons that make any such tube function.

The perfection of this 2,000,000-volt direct-current tube is the best proof we can offer of the value of the Machlett skills that go into the design and manufacture of every tube bearing our name ... Machlett Laboratories, Inc., Springdale, Connecticut.







• It's the little things that loom biggest in the manufacture of delicate electrical measuring instruments. Little things like specks of dust or breath condensation can play havoc with accuracy. That's why Triplett Instruments are made in spotless manufacturing departments; why the air is washed clean, de-humidified and

temperature-controlled; why every step in their mass production is protected. As a result Triplett Instruments perform better, last longer and render greater service value.

Extra Care in our work puts Extra Value in your Triplett Instrument.



COMMUNICATIONS FOR JANUARY 1945

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A GUIDING HAND

FLIES WITH HIM

American airmen never fly alone! Riding with them go the combined

skills and efforts of many thousands of war workers and technicians. To their targets... and safely back to their bases . . . the finest equipment that man can produce flies with them.

Eastern is proud of the part its equipment plays in these great missions. Before the war, amplifiers were thought of only for their use in sound amplification. Today, thanks to the tremendous advancement in Army and Navy practice, amplification finds new fields of service in step with vital war instruments.

Our contribution to this development has broadened the long-standing experience of Eastern engineers, who are looking ahead to your post-war needs . . . not only in sound reinforcement, but also in special amplification related to industrial instruments. But first, Eastern is applying every resource to its wartime production.

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Buy MORE War Bonds

Exhaustive DuMont guality control and life tests take all the guesswork out of CATHODE-RAY TUBE



DuMont never gambles. Thus each batch of fluorescent material is thoroughly tested in sample tubes. All new metals and other materials are likewise tested. New designs are thoroughly checked with sample production runs. The same with new processes—even to supposedly minor details such as washing and drying of glass envelopes.

DuMont life-test racks eliminate the greatest gamble of all-probable life. Percentage samples are operated in these life racks for 500 hours. A time totalizer keeps score. Tubes are tested weekly for brightness and cathode condition. And at the end of 500 hours the tubes are checked for all characteristics that might be affected by such intensive operation.

So again we repeat: there is no guesswork regarding performance and life with DuMont cathode-ray tubes.





DUMONT Quality-Control Checkup

Percentage sample tests—some 100%—are made on production runs for the following: Brilliance and screen condition—color, spots, burns, etc.

Examination of cathode by using tube as an electronic microscope. The few thousandths of an inch diameter gun aperture is magnified to $\frac{1}{2}$ inch circle to check uniformity of emitting surface. Cathode emission current is measured.

Deflection sensitivity is tested to determine amplitude of signal required to deflect beam by given amount.

Maximum electrode current. Check for excess anode current (to avoid overloading power supply of equipment using tube.)

Determining grid cutoff (when too low, tube life is shortened; too high, might exceed range of intensity control of equipment using tube).

Leakage current of various electrodes.

And other critical and vital factors entering into satisfactory, long, economical tube performance.





How MYCALEX Solved a Tough Insulating Problem for HAZELTINE ELECTRONICS and the NAVY



CLIFTON, New Jersey

NEW YORK 20, N. Y.



We can all see with the naked eye that the Payroll Savings Plan provides the most stable method of war financing. Analyze it under the X-ray of sound economics and other important advantages are evident.

A continuous check on inflation, the Payroll Savings Plan helps American Industry to build the economic stability upon which future profits depend. Billions of dollars, invested in War Bonds through this greatest of all savings plans, represent a "high level" market for postwar products. Meanwhile, putting over Payroll Savings Plans together establishes a friendlier relationship between management and labor.

To working America the Payroll Savings Plan offers many new and desirable opportunities. Through this systematic "investment in victory," homes, education for their children and nest eggs for their old age are today within the reach of millions.

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•Amphenol offers highly developed ability to design and produce electrical components of the most critically engineered type. This skill is based on many years of experience in radio since its pioneering stages-now deepened and strengthened by the extreme demands of war production.

Fortunately Amphenol's great capacity is available at a time when the whole science of Electronics is coming into its own. Amphenol is ready to play as big a part in its own phase of the Electronic Industry, as Electronics will play in the post-war world.

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POLICE COMMUNICATION PROBLEMS FACING ENGINEERS TODAY

HE postwar development and expansion of the communication services of law-enforcing organizaions presents emergency radio-communiation engineers with immediate probems more serious than any group of adio men have ever attempted to solve. The expansion of existing safety and aw-enforcement communication services nd the establishment of new services on firm economic and engineering basis vill require the highest degree of undertanding and cooperation between the egulatory branches of government and he communication divisions of the safey organizations themselves.

Radio communications in the police epartment has, from its inception over ifteen years ago, amply proved its value n the apprehension of criminals and in he coordinating of the activities of the nobile units of the police department. aw-enforcement agencies entrusted with he responsibility for public safey within

Figure 3 Captain Boss testing an h-f f-m two-way unit used by the Warwick police.

by LEWIS J. BOSS

Captain, Radio Supervisor Warwick Police Department Apponaug, R. I.

the areas protected by them must necessarily depend upon an adequate com-



munication system if they are to successfully combat crime. Without adequate and reliable means of communication their organizations and equipment are helpless and will fail to achieve their intended purpose in exactly the same degree as their communication systems are inadequate and unreliable.

Metropolitan Area Problems

Approximately 68,403,000 persons live in the metropolitan areas of the United States. In these heavily congested areas there is a vital necessity for immediate and direct communication between headquarters and those peace officers assigned to duty throughout the area. In these localities, cities and towns closely adjoin each other and their respective police departments use the radiotelephone for communication to police cars very extensively.

In the metropolitan New York area there are instances where fifteen or more police radio systems are trying to communicate on the same frequency. It is estimated that there are some 3,000 po-

(Continued from page 77)

A REPORT ON THE TECHNICAL PANEL



Highlights of Comment By M. A. Trainer, J. E Keister and W. S. Lemmo by LEWIS WINNER Editor

Walter S. Lemmon, right, and Paul Chamberlain with model of the automatic relay tower I.B.M. and G.E. propose to build for multiplex operation

THE first annual TBA conference, heid in New York City recently, saw over 1000 in attendance. Among the many interesting features offered was a technical panel meeting presided over by C. A. Priest of General Electric. Among those appearing on the panel were: Merril A. Trainer, RCA; J. E. Keister, G.E.; Walter S. Lemmon, International Business Machines; Dr. Peter C. Goldmark, CBS; Paul Chamberlain, G. E.; and David B. Smith, Philco.

Camera and Picture Tube Developments

Mr. Trainer introduced his phase of anaylsis with data on television pickup tubes that were in general use before the war. He pointed out that the most important of these have been the iconoscope, the image dissector tube and the orthicon.

He said that wherever sufficient light was available the iconoscope has been the first choice, because of its superior resolution capability and its desirable gamma characteristic. The iconoscope's light range and gamma property make it particularly suitable for scenes with ample lighting without any difficulties due to high light saturation, he stated. Thus it is quite useful for out-door pickup.

He also pointed out that the general cleanness of the picture obtained with an iconoscope had made it almost universally used as the camera tube for live talent studios. Describing resolution of the iconoscope, he said that it is sufficient at the present time to reproduce practically any picture detail which can be transmitted over the standard television channel. And, he continued, because of its *storage* property, the iconoscope has also been generally used in film cameras. Modified intermittent type projectors are used in this application, he explained. That is, the picture is flashed on the mosaic during the vertical blanking time and the mosaic is scanned during the following dark period.

Mr. Trainer also discussed some of the undesirable features of the iconoscope. There are, he said, spurious shading signals and flare which must be electrically compensated in producing a good picture. At times such compensation represents an added complication in a practical setup, he explained. In addition, he said, the relatively large size mosaic used in an iconoscope requires the use of large camera lenses. For telephoto shots this may result in relatively unwieldly cameras. Another requirement for the iconoscope is the use of keystone correction scanning, he said. This was once a more serious problem from the circuit standpoint than it is at present. He said that this draw-back might be classified more as a nuisance than a limitation.

Large orthicons are generally used for out-of-door events where the lighting is usually more unfavorable, according to Mr. Trainer. He said that its sensitivity is 4 to 5 times better than that of the iconoscope. However, he pointed out, the picture as compared to an iconoscope picture has a tendency to lack snap because of the strictly linear gamma characteristic. Accordingly, an increase of picture detail brightness by a factor two in the orthicon will increase the video signal by a factor two, explained Mr. Trainer. In the iconoscope this increase can be much less. This gamma characteristic, tending to compress whites appears to be very desirable, said Mr. Trainer. And in an orthicon camera it can be obtained by the introduction of gamma correction in the video amplifiers. but at the expense of increased noise in the picture. Mr. Trainer also pointed out that the smaller mosaic not only allows the use of shorter focal length lenses for a given field but the use of longer focal length lenses of large effective aperture for telephoto effects. The limited light range was described as one of the major and somewhat annoying limitations of the orthicon.

According to Mr. Trainer, when viewing a scene, the mosaic will charge up

with picture highlights above a rath definite value, and unless the beam cu rent is increased under these condition the charge can spread over the entimosaic, completely obliterating the piture. A spot pickup such as is encountered at a football game offers a practic example of this limitation. He said the it is very possible to have a scene which may consist of part sky and part groun or a highly illuminated section of th field together with a shadowed portio In such a case explained Mr. Traine the operator has a choice of allowing th highly illuminated portion to stick to son degree, wiping out detail in the high lights rather completely, or stoppir down the lens to the point where the highly illuminated portion is satisfactori transmitted, but generally at the expension of insufficient illumination in the log lights. The iconoscope automatical provides for this range characterist which can be so troublesome, continue Mr. Trainer.

The so-called S distortion and ion spa formation, a prewar orthicon limitatio was also analyzed. Mr. Trainer said that this distortion, has been due mainly a practical limitations of the electron-optidesign, which is relatively more con plicated than in other pickup tubes. With new developments it is probable that the distortion introduced by an orthicon a pickup of a scene can be reduced to very acceptable value.

Discussing the image dissector, M Trainer pointed out that it has not been used extensively as a live talent camer tube because of its relatively low set sitivity. However, he said, because of the absence of shading signal from the dia sector tube it has found use as a film camera tube. This is because the tub inherently does not have memory of storage. Thus it is used with a continuou film projector to produce satisfactors pictures free of shading.

Improvements in tube design were als analyzed by Mr. Trainer. He showed the the resolution capability of the iconoscop has been increased by the use of an in proved gun design to produce a better
SESSION AT THE TBA CONFERENCE

fined scanning spot. In addition, special ocessing and electrode arrangement ave helped to reduce edge flare and ading troubles. He said that a large ctor in the improved performance of e iconoscope has been our increased opreciation of the role of bias and edge chting.

The orthicon tube appears to provide e greatest possibility for radical imovement in pickup tube performance cording to Mr. Trainer. Thus far great rides have been made mainly in the rection of increased sensitivity and inge.

Covering factors which represent goals or performance, Mr. Trainer offered four octors These were:

(1)—Increased sensitivity, allowing for peration under adverse lighting condions or reduced studio lighting.

(2)—Resolution, which should be sufciently good to reproduce at least all etail which can be transmitted over asgned frequency channels. This resoluon requirement may be tempered by the erformance of a tube under low light onditions. That is, a tube which has stremely good sensitivity may have mewhat less resolution than the ideal nd still be a very useful device for pickig up spot broadcasts.

(3)—The operating light range shall c sufficient to reproduce satisfactorily dely. The tube shall be free from icking in the highlights and yet give tail in the darker portions of the pictre. The gamma characteristic should c under control so as to give desirable hite signal compression.

(4)—The geometrical distortion such s S distortion in the orthicon shall be low as to be unnoticeable on the verage transmitted scene.

icture Tubes

Mr. Trainer said that the most sigificant development in picture tubes has een the combination of a high-voltage rojection type kinescope with a Schmidtptical system to make a practical proection receiver for home use. He xplained that a 4" or 5" kinescope perating at 25 to 30 kilovolts in a schmidt-optical system having a 12" to 4" spherical mirror and a suitable corecting lens will produce a very satisfac-ory $15'' \ge 20''$ picture. Such a picture, e said, has a brightness of approximately foot-lamberts in the highlights and, at the present time, a contrast range of 5 to 1. He pointed out that experiments with many types of translucent projection creens used with this system indicate hat it may be possible to increase the ontrast ratio to approximately 40 to 1, which is about as good as can be obtained n a direct viewing kinescope. Describing he Schmidt optical system, he said it has In equivalent f number of 0.9 and a hrow of 34" to 36"; capable of 1500-

ine resolution and should therefore inproduce no loss in detail with a 525-line incluminized glass spherical mirror with molded-plastic correcting lens repreprenented the most satisfactory Schmidt sys-



Typical setup of equipment that will be employed by satellite stations, according to J. E. Keister of General Electric.

tem combination for projection receivers at the present time.

Postwar direct-viewing picture tube improvement trends include operation of the kinescope with 12,000 volts or more on the second anode instead of the 6000 volt rating used before the war, according to Mr. Trainer. This voltage provides a substantially brighter picture. Horizontal back-kick rectifier systems will probably provide the necessary d-c voltage, he said. Flat-faced tubes appear to be another trend. Two types which give promise of being widely used according to Mr. Trainer, are the 10" flatface round tube and the 14" diagonal flat-face rectangular tube. Mr. Trainer said that the use of these tubes will allow for a more pleasing picture of larger size than was practical with tubes of prewar design.

Ion spot formation common to prewar kinescopes, will probably be greatly retarded or eliminated completely in postwar kinescopes, disclosed Mr. Trainer.

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SATELLITE TELEVISION STATIONS

THE place of satellite stations in a television system served as a basis

of a discussion by J. E. Keister. He said that a satellite station in its simplest form is one which originates no local live talent programs, serving its locality by broadcasting programs fed to it by a network. In some localities satellite stations may be installed by the television network operators themselves to serve the smaller communities he said. Thus, he explained, stations serving smaller communities can supply the quality of entertainment or educational programs which otherwise could only be justified by a few installations in densely populated areas.

Mr. Keister explained that most relay systems contemplated at present expect to supply at least two channels in each direction on the main artery. Thus two sources of programs can be originated simultaneously on a single network setup. A satellite station then has a choice of programs with the accompanying benefit of variety to fit its own particular needs, he said.

It usually can be assumed that the satellite station is off the main artery and so a simple one-way branch relay will be used, according to Mr. Keister. In many cases a single jump from the main relay tower to the satellite station will suffice, he explained. If not, simple one-way repeater stations would be installed at the required intervals.

Describing the satellite station itself, he said that it consists essentially of a receiving antenna to pick up the relayed signal, receiver for amplifying and demodulating both the aural and visual components, monitor to provide a running check on the signal, and finally the main transmitter for broadcasting the sound and picture to the surrounding area. The actual power output will depend on the topography and area served, he said. Since the FCC in authorizing the construction of satellite stations doubtless will require some means of both aural and visual station identification, equipment for such identification will be necessary, Mr. Keister stated. This would involve the inclusion of a monoscope signal generator, as associated pulse generator and sweeps, and suitable audio and transcription equipment; personallized monoscopes can be anticipated with the stations own call letters inscribed.

Since the interest shown in local news events will usually justify the addition of equipment for the origination of film programs, many satellites will require a motion picture projector, film pick-up camera, camera amplifier and control equipment. With these film origination facilities available, said Mr. Keister, the local satellite station operator can, with a 16-mm silent motion picture camera, take pictures of local events.

By capitalizing on this simplified type of station to augment the stations serving (Continued on page 75)



Radiotelegraph pack set used by the infantry.

Equipment shown in these British Official photos was designed for airborne operations, such as those recently carried out in Holland. (Crown Copyright Reserved)

Right, a short-range radiotelegraph unit used by the infantry.









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Left, above: A me-

dium-range unit for

telephony and teleg-

raphy used in vehi-

cles and by ground stations. Left, below: Medium range radiotelephone set for tanks or trucks. (*Truck view, courtesy U. S. Signal Corps*) Above, a medium-range radiotelephony set similar to that shown mounted in comman car at lower left. Below, the telephony ar telegraphy unit, shown at left above, in container for dropping by parachute.



AIRBORNE EQUIPMEI

when to Court 181 Stations DEVOTED TO RESEARCH AND THE MANUFACTURE OF TUBES AND EQUIPMENT FOR THE NEW ERA OF ELECTRONICS RATTHEON MANUFACTURING ENTIRE BLUE NETWORN Waltham and Newton, Massacl COMPANY ELECTRONIC AND RADIO TUBES High Fidelity RAYH 3,500,000 Radio Homes Each 100 16 11/16 Now Garri Ask your Raytheon distributor for colorful, attention-getting "Meet Your Navy" display to tie in with Two Stars his great program. Easel-mounted, 171/2 inches x 201/2 inches. JANUARY 1945 37 .







Figures 1, 2 and 3 (left to right, above)

Figure 1, a single-tuned circuit with a series impedance z. Figure 2, the single-tuned circuit with a parallel impedance Z. Figure 3, circuit with a resistor across to reduce effective Q, providing a resonant impedance Z'o.



by J. E. MAYNARD

Transmitter Division **Electronics Department General Electric Company**

NE of the subjects treated in some detail in the ESMWT RPI course was that of tuned and coupled cir-It may be shown that a single tuned circuit has the series impedance z, Figure 1,

$$z = r (1 + jS)$$

and has the parallel impedance Z_{i} Figure 2,

$$Z = Z_{o}/(1 + jS)$$

 $S = Q_o (\Delta f/f_o) (1 + f_o/f)$

in which zero subscripts refer to values



at resonant frequency f_o , where, by definition, $X_L = X_C = X_o$. Frequency deviation $(f - f_0)$ is represented by Δf_i , and Qo is the quality factor of the circuit at f_o. The impedance Z_o is parallel impedance at resonance, $Z_o = X_o Q_o$. Equation 2 is a very close approxima-tion if Q_{\circ} is 10 or more, which covers essentially all coils used in tuned cir-



(1)

(2)

COMMUNICATIONS FOR JANUARY 1945

If Δf is 10% of fo or less we can cuits. make a further simplification by the approximation

$$S = (2\Delta i / f_{e}) Q_{e}$$
(3)

Now if we place a resistor across the circuit, which is often done to reduce the effective Q of the circuit, its effect may be accounted for by considering the circuit to be one having a new resonant impedance Z'.

$$Z'_{o} = Z_{o} R_{o} / (Z_{o} + R_{o})$$
⁽⁴⁾

Since $Z_0 = X_0 Q_0$, the new circuit has a new quality factor

$$Q'_{o} = Q_{o}R_{o}/(Z_{o} + R_{o})$$
⁽⁵⁾

The reactance X_o is, of course, the same in either case under the approximations we have made.

Coupled Circuits

Any pair of tuned circuits may be coupled in several ways, Figure 4. Pi networks may always be reduced to equivalent T networks² and circuits coupled by inductive fields may always be reduced to an equivalent T of inductive coupling, although it may require a fictitious negative inductance for reversal of polarity.

Now if we disregard the case in which capacitive and inductive coupling become equal and opposite, a possibility for one polarity in circuit (f), then all these cases may be represented by one general circuit, Figure 5.

This represents a primary resonant circuit of series impedance z, coupled to a secondary resonant circuit of series impedance z_2 by a common impedance z_m It will be assumed that zm contains no resistive components and further that its value remains fixed at its resonant frequency reactance. The result of these assumptions is symmetry about fo. This conforms closely to practical performance when Δf is less than 10% of f_o , and when the coupling impedance has a nor mal quality factor.

The circuit of Figure 5 may be solved for either i, or is in terms of the driving voltage e. Starting from the circuit equations

$$e = i_1 z_1 - i_2 z_m$$
 $0 = i_2 z_2 - i_1 z_m$ (6)

Figure 4

Various methods used to couple circuits.



(7)

e obtain

$$= i_1(z_1 - z_m^2/z_2)$$

$$(z_m/z_1) = i_2(z_2 - z_m^2/z_1)$$
 (8)

These solutions illustrate a concept of e circuit operation which is often conenient. The impedance in the primary ad also in the secondary has been panged by a *coupled-in* impedance z_m^2/z_a the primary and z_m^2/z_1 in the secndary. The driving voltage in the secndary may be considered as the driving oltage e divided in the ratio z_m/z_1 . hese two equations are the funda-entals of a pair of coupled circuits. rom equation 8 we can determine what appens to a signal as it passes through; om equation 7 we can determine what rt of impedance the primary presents the driving circuit.

oupled Circuit Selectivity

The values of z_1 and z_2 are the series npedances r(1 + jS) for the primary id secondary respectively. The common ipedance z_m becomes $+jX_m$ for in-ictive coupling or $-jX_m$ for capacitive upling. Using the coefficient of coupng $K = X_m / \sqrt{X_1 X_2}$ and defining the llowing new terms

geometric mean $Q = Q_g = \sqrt{Q_1 Q_2}$ arithmetic mean $Q = Q_a = (Q_1 + Q_2)/2$ equivalent equal $Q = Q = Q_g (Q_g/Q_a)$ selective variable $S = (Q/Q_1)S_1 = (Q/Q_2)S_2$ oportional coupling C, $1 + C^2 = (Q_g/Q_a)^2 + K^2Q^2$ attenuation $U = (gain at f_0)/(gain at f)$

may be shown¹ that the attenuation or lectivity curve for a pair of coupled #rcuits is

$$J = \frac{\sqrt{(1 + C^2 - S^2)^2 + 4S^2}}{1 + C^2}$$
(15)

his follows from equation 8 and is a eneral expression for attenuation which ay be utilized for circuits having the me or different Q in primary and secidary. The proportional coupling C is a value of 1.0 for a flat charachdary. (Fristic at the transition from single to buble peaks.

To reduce Figure 5 to a practical sysm, let us consider the circuit of Figte 6. The tube is now providing the riving voltage e of Figure 5, all circuit ading is lumped in the resistors R_1 and

RCUIT DEVELOPMENT



Figure 5

 R_2 so that $Q_{\circ 1}=R_1/X_{\circ 1}$ and $Q_{\circ 2}=R_2/X_{\circ 2}$. It is, however, notable that the tube is driving across the primary, whereas e in Figure 5 is a series driving The two situations may be voltage. reconciled. A tube driving a load may be represented as a voltage source or as a current source, Figure 7.

The voltage $e_{\mathbf{r}}$ developed across Z must be the same by either representation.

$$\mu e_{g}Z/(R_{p}+Z) = g_{m}e_{g}[ZR_{p}/(Z+R_{p})]$$
(16)

Since $g_m = \mu/R_P$ the identity of results is apparent. Figure 6 may therefore be represented by a current gmeg operating into the primary with R_p absorbed in R₁. Now we can relate Figures 5 and 6 through Figure 8. Circuit (a) represents Figure 5 and circuit (b) represents Figure 6.

The relation to be established is that the current i be identical in the two cases. This means that the voltage drop $g_m e_g Z_1$ must equal the driving voltage e. The impedance Z_1 may represent the re-actance of C_1 , and Z_x the impedance of the remainder of the circuit including R1 and the coupled-in impedance from the Z1, of course, varies with secondary.

(9)	•
(10)	Figure 6 Figure 5 reduced to a practical system, with
(11) (12)	the tube providing the driving voltage e and all circuit loading
(13)	lumped in resistors R ₁ and R ₂ .

frequency; this produces non-symmetry which will be neglected so that e may be assumed constant with frequency variation for practical calculations

$$e = -jg_m e_g X_{o1} \tag{17}$$

Primary Impedance

If we are interested in the plate voltage appearing at the primary circuit either for determining overload conditions or for determining performance of such circuits as discriminators, avc rectifiers, etc., we will want to know the impedance at the primary from plate to Suppose, therefore we identify ground. the a-c plate current in Figure 6 as i, and a-c plate voltage as ep, then primary impedance $Z_p = e_p/i_p$. In equation 7 the driving voltage $e = -jX_{olip}$ and the primary circuit current $i_1 = e_p/-jX_{o1}$. From these relations we find

$$Z_{p} = -X_{o1}^{2}(i_{1}/e)$$
⁽¹⁸⁾

or, using equation 7

$$Z_{\rm p} = -X_{\rm o1}^2 z_2 / (z_1 z_2 - z_{\rm m}^2)$$
(19)

From equation 8

$$z_m(e/i_2) = z_1 z_2 - z_m^2$$
 (20)

Secondary voltage $e_2 = -j X_{02}i_2$. Substituting for e and i_2

$$-X_{o1}X_{o2}(i_p/e_2)z_m = z_1z_2 - z_m^2$$
(21)

Using $i_p = g_m e_g$ and substituting for $z_1 z_2 - z_m^2$ in equation 19

$$Z_{p} = (X_{o1}/X_{o2}) (z_{2}/z_{m}) (e_{2}/g_{m}e_{g}) \qquad (22)$$

From equation 22 we can anticipate the shape of the primary impedance curve with frequency variation. Thinking in terms of a constant applied grid voltage eg, the secondary voltage e2 will trace out the familiar coupled circuit response. We may assume everything constant except e_2 and z_2 and z_2 is the series resonant characteristic of the secondary circuit, $r_2(1 + j S_2)$. The z_2 multiplier serves to produce a dip in the characteristic at resonance. Physically the secondary is acting as an absorption circuit, reducing primary impedance at resonance. The primary impedance will always have a deeper dip at resonance than the overall response curve.

The attenuation on a selectivity curve, U, has been defined as the ratio of gain at resonance, A_0 , to gain at another fre-quency, A_r . Since e_2/e_g is gain at any frequency, it may be replaced by A_0/U . A_o may be obtained from equation 21.





Figure 9 A form of the Foster-Seeley discriminator circuit.



Figure 10

Fundamental i-f circuit involved in Figure 9, redrawn as a bridge and omitting *M*. In (a), primary is connected to center of secondary coil; (b), primary is connected to center of secondary capacitor.



Figure 11 Resultant vector voltage relations.

In plotting this



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The discriminator S curve.

Figure 12

curve, the following data and Figure 11 were used: The voltage e' is rectified to produce a

positive d-c voltage E', and e'' is rectified to

produce a negative d-c voltage E''. The resultant voltage E is the algebraic sum of E' and E''.

And at the larger frequency deviations, the se-

lectivity of the circuits reduces all voltages.



where $Z_{og} = X_{og}Q_g$, $X_{og}^2 = X_{o1}X_{o2}$. From equations 24 and 11 it can be deduced that the greater the difference between secondary and primary Q, the greater i the bandwidth (less equivalent equal Q) that will be obtained for a given amoun of gain.

Equation 22 may be rewritten with th aid of equation 24 and some algebrai manipulation

$$Z_{p} = [Z_{o1}/(1 + K^{2}Q_{g}^{2})] [(1+jS_{2})/U]$$
(25)

The expression for U given in equation 15 omits phase; if phase relations are retained, it may be shown that

$$U = (1 + C^2 - S^2 + 2jS)/(1 + C^2) \quad (26)$$

With this information we can readily calculate the impedance in phase an magnitude into which a tube works with any pair of coupled circuits at its plate This impedance may then be plotted of the tube characteristics as a load line o ellipse to determine maximum power out put and overloading point at any free quency.

Discriminator Circuit

One form of the Foster-Seeley dis criminator³ is shown in Figure 9. L i a choke coil which provides a d-c retur path for the diodes and C_b is in each case a blocking or a bypass capaciton The coupling in this circuit is shown ad existing in the inductive field betwee primary and secondary. Capacitive coup ling is also possible in a discriminato circuit. In fact both types of couplin will ordinarily exist simultaneously in circuit such as figure 9. To understan the nature of the capacitive coupling is best to view the circuit as a bridge The fundamental i-f circuit involved omitting M, is redrawn as a bridge i Figure 10. Two possibilities are shown in circuit (a) the primary is connecte to the center of the secondary coil, whil in circuit (b) it is connected to the cer ter of the secondary capacitor. The driv ing voltage for the bridge is e_p and i output voltage is e_2 . C'_a and C''_a are the transformation of the contrast stray capacities to ground from the plate of the two diodes. If the arms of the bridge are perfectly balanced no secondar voltage is produced and we have zer coupling. If the center of the coil or ca pacitor is not electrically true, we have inductive or capacitive coupling respe tively. If C'a and C"a are unbalanced w have capacitive coupling. Any mutu coupling between coils will, of cours add algebraically with coupling due

CIRCUIT DEVELOPMEN

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idge unbalance. Coupling by bridge unllance can often be used to advantage, c., for obtaining a discriminator characristic from such devices as crystals⁴.

Iscriminator Analysis

The Foster-Seeley discriminator transtimer develops the S curve of voltage ainst frequency by rectifying, in opsition, voltages resulting from the ctor addition of the primary and secdary voltages in the coupled circuits. from Figures 9 and 10*a* the vector a-c ltages from the diode plates to ground e seen to be

$$= e_p + e_2/2$$
 (27)

(28)

$$= e_n - e_2/2$$

r one polarity of coupling; signs will reversed for a reversal of coupling. hese voltages must, of course, be added ctorially. We must, therefore, deterine the vector primary and secondary pltages. The primary voltage is a-c ate current $(g_m e_g)$ multiplied by imary impedance, Z_p . Using equations i and 26, and letting $1 + jS_2 = U_2$, we ave

$$= [g_m e_g Z_{o1} / (1 + K^2 Q_g^2)] (U_2 / U) \quad (29)$$

ince $A_r = A_o/U$, equation 24 may be sed to determine the vector voltage e_2 .

$$= (jg_m e_g Z_{og}/U) [KQ_g/(1+K^2Q_g^2)]$$
(30)

The variations in phase positions of nese voltages are dependent on U_2 and 7. At resonance (S = 0), U_2 and Ure both real and unity so that primary nd secondary voltages are in quadrature. The ratio of magnitudes and the phase ngle between the two voltages can be btained from the quotient of the vectors.

$$_{\rm p}/{\bf e}_2 = j \left({\rm U}_2/{\rm KQ}_{\rm s} \right) \sqrt{Z_{\rm o1}/Z_{\rm o2}}$$
 (31)

Figures 16 (below) and 17 (right) loaded i-f circuit representing the discrimiator circuit of Figure 9. The effect of the iode loads is incorporated in the fictitious reistors $\frac{1}{2}R$. Figure 17, a discriminator curve plot.

IRCUIT DEVELOPMENT

• Figure 14 A diode rectifier circuit where the peak d-c voltage developed is 1.414 E; the i-f volt-

age driving the diode is represented by an rms value E.

Since $U_2 = 1 + jS_2$ the phase angle between e_p and e_2 is

(32)

 $\phi = \tan^{-1}S_2 + 90^\circ$

.

Figure 15

Another diode rectifier circuit. Here, however, with the same R-Ctime constant the performance will be the same as that of Figure

14, except for the appearance of i-f, as well as the d-c voltage across R.

The resultant vector voltage relations are shown in Figure 11.

Discriminator Equations

A method of calculation may be obtained from Figure 11.

$$(e')^{2} = e_{p}^{2} + (e_{2}/2)^{2} + e_{p}e_{2}\cos\phi$$
 (33)

The voltage e' is rectified to produce a positive d-c voltage E', and e" is rectified to produce a negative d-c voltage E". The resultant voltage E is the algebraic sum of E' and E". At the larger frequency deviations, the selectivity of the circuits reduces all voltages. From these considerations and Figure 11 the manner in which the discriminator S curve is obtained will be apparent, Fig-

$$\cos \phi = -S_2/U_2 \tag{36}$$

And we have

(Continued on page 68)

Figure 1*a* Extreme left, an external anodetype tube (GL-880) made by G.E.

Figures 1b and c Center, RCA 827-R; and right, Westinghouse WL 895R, external anode-type tubes.

EXTERNAL ANODE TRIODES CHARACTERISTICS AND APPLICATIONS

[PART I OF A FOUR-PART PAPER]

by A. JAMES EBEL

Chief Engineer WILL, University of Illinois

E ARLY in the development of highpower radiotelephone transmitters it became apparent that vacuum tubes cooled by direct radiation could never fulfill the dissipation requirements imposed by operation in the kilowatt-power ranges. First, the dissipation of each tube was limited by certain physical factors in the tube design; and second, the operation of a number of tubes in parallel led to the well-known parasitic instabilities of such circuits. A solution to the problem was found in the external anode type of tube. The perfection of the metal-to-glass seal paved the way for the development of these high-dissipation power tubes.

Today there are many types of external-anode vacuum tubes varying greatly in size and design. Some are cooled by circulating water, and others by forced air. The dissipation for the various types ranges from 1.2 kilowatts to 250 kilowatts. The smallest can be held in the palm of the hand, whereas the largest stands nearly as tall as a man. A representative cross section of these tubes is shown in Figure 1a, b, c, d, e and f. Until recently such tubes were used principally in radio transmitters. But the recent introduction of high-frequency heating technique into many industrial operations has opened up a new field of application that promises to surpass the utilization in radio. It is important that the operators of equipment using external anode tubes understand fully their characteristics, their proper care, and something of their general construction. The purpose of this paper is to offer such information.

With the scarcity of materials and manufacturing facilities during war comes the necessity for increased attention to any details which tend to prolong the life

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Figures 2 (above) and 3 (below). Figure 2, six-strand filament of a high-power triode. (Courtesy G.E.) Figure 3, detailed view of high-powered triode grid structure. (Courtesy FTR.)

of such tubes. A wealth of valuable material has been published^{1, 2, 3} on this phase of tube operation and will be referred to, in this paper, from time to time.

The filaments of external-anode tubes are composed of pure tungsten. In its pure form this element has the highest melting point of any ductile metal (3655° K) and emits electrons in satisfactory quantities when operated at temperatures at which exaporation is not impractically great. Figure 2 shows the construction of a six-strand filament for a high-power triode. It will be noted that leads are brought out from each of the strands so that the filament may be connected for single-, three-, or six-phase operation. The pure metal becomes ductile when properly worked and can be drawn into strands of the proper size. After being heated to incandescence, the metal returns to a crystalline form as it cools, and becomes very brittle. To prevent these brittle strands from shattering, they are mounted with springs at either end to take up the slack caused by expansion when the filament is heated and to protect the strands from physical shock. One of the fundamental necessities in the care of tungsten emitters is to protect them from all shocks because they are so easily fractured. At least one leading transmitter maintenance authority⁴ recommends that the tubes not be removed from their sockets during their entire life. Other considerations bearing on the advisability of this procedure will be mentioned later.

The grid structure is generally a helix. wound around several supporting rods so that the spacing between the grid wires and the geometrical configuration of the filament strands are uniform. The grid operates at a fairly high temperature and is called upon to dissipate considerable power. The grid size is a compromise between a thick strand, which is best for heat dissipation purposes, and a fine strand, which has the best electrical characteristics. The material must have a high melting point and a high work function for mechanical stability and low secondary emission. Such metals as tungsten, tantelum, and molybdenum are normally used for both the grid wires and the supports of these wires. Special consideration must be given to the grid lead to provide high insulation, low capacity to other elements, and low I2R loss. In Figure 3 we have a detailed picture of the grid structure of a high-power triode. The supports for the grid and filament structure must be carefully machined to provide accurate spacing and to maintain that spacing. Any warping will cause

a change in tube characteristics and may lead to interelectrode shorts. In the case of some ultrahigh-frequency tubes this spacing is as low as .06", indicating the necessity for carefully maintaining the correct alignment.

Since it is particularly important that the copper used be oxygen-free, the anodes of these tubes are made of the purest copper available. The metal used in each tube must be from stock that has been carefully analyzed for foreign materials and gases, and throughout the manufacturing process careful checks must be made to see that no gas, especially oxygen, has been introduced into the copper. Pure copper is necessary to effect a good metal-to-glass seal, to give proper strength, and to maintain the vacuum within the tube. For a successful metal-to-glass seal the glass must wet the metal; that is to say, it must flow. over the metal rather than collect in a globule as water does on a greasy surface. Furthermore, it is necessary to provide for the differential expansion of the metal and the glass. Figure 4 shows the expansion of the normally used metals and glasses as related to temperature changes. Note that the expansion of glass and copper are decidedly different. In spite of this differential, the high ductility and the low yield point of copper make possible the successful combination of these two materials. It is necessary, however, that the copper be thin at the junction, a fact which tends to produce mechanical weakness in the tube. This is another reason for careful handling of the tube. The seal of the external anode tube is mechanically its weakest point and must be treated as such. In a typical po 5-kw tube the anode wall thickness is .065". This thinness is required almost This thinness is required almost entirely by thermal considerations.

High temperature is destructive to the was seal. Fortunately there is generally a noticeable discoloration of the seal as the temperature becomes too high. This is especially noticeable in the copper-toglass seals where the change is from the normal reddish copper hue through green to a dark metallic color.

As has been stated, there are two methods of cooling the anodes of these tubes, water cooling and forced-air cool ing. The water-cooled tube is mounted in a jacket with provision for water entering the bottom of the jacket and leaving the top. Since the anode of the tube operates at a high d-c potential, it is necessary to provide adequate insulation in the water circulating system be-(Continued on page 104)

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CAA checker-board radio-range frequency assignment plan dividing the country into 100-mile squares with one range station within each square

AERONAUTICAL COMMUNICATIONS

Consulting Engineer

F ROM War I up to about 1940 aeronautical radio communications seems to have "just growed." It can hardly be said, since no one could prophesy the extent of character of aviation as of 1940 at any significantly earlier date, that any particularly intelligent planning for the future could have been engaged in. Thus the assignment of aircraft communication and navigation frequencies, establishment of policies for regulation of aeronautical radio communication and a host of other details had, of necessity, to await the development of a sufficiently clear-cut aviation situation.

By approximately 1940 the aviation communications and navigation outlook for at least the immediate future had crystallized sufficiently to permit of intelligent long-range planning. The results of such government planning were partially set forth in FCC Rules and Regulations Governing Aviation Services, (Title 47—Telecommunication—Chapter I, Part G), effective August 1, 1939 and revised to November 1, 1942. This publication, available from the Superintendent of Documents, Washington, D. C., at a price of five cents, together with a copy of the National Allocation Plan For Assigning Radio-Range Frequencies in the Band 119-126 Megacycles which was printed by CAA as Technical Development Report No. 28 in May, 1941, and which was prepared by L. H. Simson of CAA constitute must information for anyone seriously interested in the probable extensive expansion of aeronautical communications after V-Day.

Desiring not to explore in detail the data set forth in referenced publications, but rather to consider briefly a number of aspects of *postwar* aeronautical radio communications the writer will pass lightly over the postwar frequency situation, particularly as, while in principle it will probably conform to the excellent plan already devised, it is possible that actual frequencies may be somewhat different.

In the prewar period government and commercial aeronautical radio services operated upon a number of different bands. Possibly most familiar is the 200to 400-kc low-frequency band in which were located radio-range, communication and navigation services. In this band 278 kc was uniformly assigned to a port control stations for communicati with civilian, itinerant and other aircr in flight. Such aircraft were expected call and work the airport within rar on 3105 kc in daylight or 6210 after dark. In practice, 6210 kc was lit used, probably because 3105 kc provid sufficient distance range even after da to satisfy the limited-range flight of 1 average civilian flyer—and it is with h and his brothers-legion that the writer primarily concerned.

Experience having indicated these 1 and medium frequencies to offer int ference possibilities incompatible with future concept of one or more airpo in almost every community, and the sands of civilian-itinerant aircraft flight simultaneously, the *u-h-f allocat plan* was wisely devised. Under this p it was proposed to shift all such ae nautical radio communication and navi tion services up into the range of to 156 megacycles—a far cry indeed fr the older low and medium frequenc which were to be progressively abandor Such a shift was most logical and s sible, for frequencies within the n (Continued on page 76)

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Left, mixer attenuator. (Courtesy Cinema Engineering Company) Above, precision attenuator box. (Courtesy General Radio Company)

RESISTIVE ATTENUATOR, PAD AND NETWORK Theory and Design

In part 11, the fundamental method of procedure in the design of the various types of attenuating networks was briefly outlined and a few of the more elementary forms of networks were considered. The second set of Tables of Hyperbolic Functions of a Real Variable was also presented, accompanied by charts showing in compact form all of the information needed for the use and design of the simpler types of networks. The series, shunt, series and shunt, and the L-type taper pads were described. The two charts also contained formulas, in explicit form, and are particularly applicable to the tables presented in the first three parts of this series.

In this installment, part III, two additional charts are shown with explicit formulas for the complete design of the T, π , bridged-T, lattice, and the multiple-bridge types of networks. These charts, like those in part II have been designed for specific use in connection with the Tables of Hyperbolic Functions of a Real Variable. The first two sets of these tables were presented in parts I and II. The third set of tables appears with this installment. The design of attenuators proceeds with a continuation of the L-taper networks, followed by the T-type network.

Appendix 1 shows the hyperbolic formulas which were used for the calculation of all of the Tables of Hyperbolic Functions of a Real Variable presented in this series of papers.

Appendix II describes a useful mathematical tool, little known by radio or communication engineers, the Gudermannian Function, which may be used for the calculation and interpolation

[PART THREE OF A FOUR-PART PAPER]

by PAUL B. WRIGHT Communications Research Engineer

of hyperbolic functions of a real variable. When any one of these is known, all of the remaining functions may be found by direct application of common tables of the natural trigonometric functions from circular function theory.

T is convenient for design purposes to utilize the unbalanced form of network configuration on a unit basis for

T-pad, 600-600 ohms; 20 steps, 2 db per step. (Courtesy Tech Laboratories)

the T, π , and bridged-T structures. These configurations are shown in the accompanying charts, as well as those of the lattice and multiple bridge types of networks. Complete design information is contained in these charts to enable the engineer to obtain the transmission loss incurred by their use in any system when operated between image impedances having the values given in the charts. The network elements are presented in symbolical form so that the equations may be solved in the most direct manner through use of the Tables of Hyperbolic Functions of a Real Variable presented in parts I, II and this installment. The definitions and requirements which must be met for the equations to be valid appeared in charts accompanying part II and this presentation, also.

It should be noted that the impedance ratio shown in the charts and development of equation is $s^2 = Z/z$. The ratio of the terminating image impedances can of course be replaced by an ideal trans-former having an impedance transforma-tion ratio of s^2 to 1. This procedure places the network proper upon an unconditionally unit basis, with the high side of the transformer terminated by an impedance of s² ohms, and gives unit or one-ohm terminations to the network. The reduction of a network to a unit basis is known as normalizing. However, throughout this paper, the ideal trans-former will be omitted and the more direct approach of the actual terminations used for the network will be used. The image impedance theory usage results in considerable simplification of the final form of the derived equations. The image propagation function, Θ , is sometimes referred to as the image transfer constant. Numerically, it is equal to

... in Directional Microphones

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NETWORK CONFIGURATION	TYPE	IMPEDANCE BASIS	DECIBELS TRANSMISSION LOSS 20 Log ₁₀ k	IMAGE IMPEDANCES	NET WORK ELEMEN TS	DEFINITIONS AND REQUIREMENTS
p [™] j [™] y z≹ ≹w ₹,	т	FULL MATCHED	k = s(I+ V+Z)	$Z = (\Phi \beta)^{\frac{1}{2}}$ $z = (\Phi \beta)^{\frac{1}{2}}$	u = cZ- w v = cz - w w = ay	$\begin{array}{llllllllllllllllllllllllllllllllllll$
		IMAGE IMPEDANCES	k = (1+ <u>v+z</u>)	z = [u(u+2w)] ^{1/2}	u = v = Dz w = az	Z = z
	т	UNIT MATCHED	k = s(l+ <mark>v∔l</mark>)	$s^{2} = (\mathbf{\Phi}' \mathbf{\beta}')^{\frac{1}{2}}$ $I = (\mathbf{\Phi}' \mathbf{\beta}')^{\frac{1}{2}}$	u' = c s²- w v' = c - w' w* = a s	$ \begin{array}{llllllllllllllllllllllllllllllllllll$
		IMAGE IMPEDANCES	k = (I + u'+I)	I = [u'(u'+2w')] ^{1/2}	u'= v'= D w'= a	Z = z s ² = a = csch o D = tanh 2 u'= u/z v'= v/z w' = w/z
	π	FULL MATCHED	k = (I+₩+ <u>₩</u>)/s	$Z = u(\sigma \rho)^{\frac{1}{2}}$ $z = w(\sigma / \rho)^{\frac{1}{2}}$	u = y²/(cz-ay) v = Ay w = y²/(cZ-ay)	$Z \ge z \qquad s^2 = Z/z \ge 1 \qquad y^{2_2} Z z$ $A = sinh_{\Theta} c = coth_{\Theta}$ $\sigma^{\Psi} v/(u + v + w) \qquad \rho = \frac{v + w}{u + v}$
) (~~	IMAGE IMPEDANCES	$k = (1 + \frac{v}{w} + \frac{v}{z})$	z = <u>uv</u> [v(v+2u)] ¹ /2	u = w=dz v = Az	Z = z s ² = 1 A = sinh⊕ d = coth 2 ⊕ = 0.115129 × No (db)
<u>ب/z</u> ۲	π	UNIT MATCHED IMAGE IMPEDANCES	k = (1+v'+ <mark>v</mark> .)	s² u'(ơ',ρ') ⁱ ² l = w'(ơ',/p') ⁱ ²	v' = A s w' = s/(cs - a)	$Z \ge z \qquad s^{2z} = Z/z \qquad A = \sinh \Theta$ $u' = u/z \qquad \rho' = \frac{v'+w'}{u'+v'} \qquad c = \coth \Theta$ $w' = w/z \qquad \sigma' = \frac{v'+w'}{v'+w'}$
4	77		$\mathbf{k} = (1 + \mathbf{v}' + \frac{\mathbf{v}'}{\mathbf{u}})$	$I = \frac{u v'}{[v'(v'+2u')]^{\frac{1}{2}}}$	u = w = d v	Z = z s² = 1 A = sinh o u' = u/z d = coth § v' = v/z w' = w/z
	MULTIPLE	FULL MATCHED	k = n _m s	INFINITE	R = [cZ - a y] r = [cz - a y]	Z≥z s ^e = Z/z≥1 y ² =Zz a=csch⊕ c=coth⊕ n=No.of branches in multiple
	MULTIPLE BRIDGE	IMAGE IMPEDANCES	k = n _m	INFINITE	R≒r=ἑDz	Z = z s² = Z/z = 1 D = tanh 🖗
	MULTIPLE	UNIT MATCHED	k = n _m s	INFINITE	R ⁴ =½[cs²~as] r ⁴ =½[c − a s]	Z≥ z s² = Z/z ≥ l c = coth⊕ a = csch⊕ n_m No. of branches in multiple
R/2	BRIDGE	IMAGE IMPEDANCES	k = n _m	INFINITE	R'= r'= ½D	Z = z s [≇] = í D = tanh [⊕] R'= R/z r'= r/z

 $0.115129 \times \text{No.}$ (db) loss of the network. It is also equal to one-half the natural logarithm of the ratio of the power delivered to the network to the power delivered to the load. This ratio has been designated a factor, k^2 .

Charts

The charts which have been prepared giving detailed design information for all of the standard as well as non-standard attenuating types of resistive networks were especially designed for direct application of the *Tables of Functions* that are described in Appendix *I*.

In the application of the formulas involving networks placed upon a unit basis, the use of primes or subscripts is undersirable and unnecessary, but to avoid some confusion which might arise in transcribing formulas from the chart, the subscript or the prime has been used. In practice, the network would be designed on the unit basis with simple regard to designation, then the multiplying factor, z, would be applied to all elements and terminations of the network to place it on a full impedance basis.

To reduce the unwieldly form of some expressions to a form more suitable for writing and manipulation, some compression has been made possible by the use of Greek symbols as well as English symbols. This was necessary to avoid duplication of the meaning of English characters and the pitfalls resulting from placing various subscripts with the same

Figures I to 6

The transmission losses and element values for the *T*, *Pi* and *Multiple Bridge* networks in terms of the *Hyperbolic Functions*.

characters to designate different things. This method of writing expressions frequently permits a minimum of operations to be performed with the equations of the network, completely devoid of the cumbersome algebraic expressions which frequently obscure the objective. As examples of this method, by inspection, one may write for the product and quotient of the image impedances of the T network, from Figure 1, $Zz = \varphi = uv + uw + vw$, and $Z/z = \beta = (u + w)/(v+w)$; and for the π network, of Figure 3, Zz = $uw\sigma = uvw/(u+v+w)$, and Z/z =uv/w = u(v+w)/(w(u+v)). Other useful relationships may be utilized to quickly check the accuracy of the calculations of the network elements obtained from the tables. An example of this may be shown from Figure 7, where $t = D_z$ and u = dz; hence, $tu = z^2$, since Dd = I, or from Figure 3,

$$\frac{1}{w} - \frac{1}{u} = c \left(\frac{1}{z} - \frac{1}{z}\right), \text{ etc.}$$

T Network

Referring to Figure 2 and the nota-

tion used for the element values, t image impedances on a unit basis may written,

 $s^2 = u' + \frac{w'(v'+1)}{v'+w'+1}$

and

1

$$= v' + \frac{w'(u' + s^{a})}{u' + w' + s^{a}}$$

The power delivered to the network free a constant current generator produci unit current and having an internal in pedance of Z ohms is

$$W = i^2 Z = Z$$

The power delivered from the network output terminals to the load impedance $w = i_z^2 z = (w'/(v' + w' + 1))^2 z$ (From expressions 3 and 4 the power ratio

W/w =
$$k^2 = ((v' + w' + 1)/w)^2 s^2$$

where
 $\epsilon \theta = k \ge 1$

$$s^2 = Z/z \ge 1$$

Since the loss in

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NETWORK CONFIGURATION	TYPE	IMPEDANCE BASIS	DECIBELS TRANSMISSION LOSS = 20 Log ₁₀ k	IMAGE IMPEDANCES	NETWORK ELEMENTS	DEFINITIONS AND. REQUIREMENTS
Lung win		FULL MATCHED	$k = s \left[\frac{z \psi * x_{z} y_{z}}{u v - 1 w} \right]$	$Z = \left[t \cup \frac{x_i y_i}{x_z y_z} \right]^{\frac{1}{2}}$ $z = \left[t \cup \frac{x_z y_z}{x_i y_i} \right]^{\frac{1}{2}}$	$ \begin{array}{l} t = \left[y(s + cx) / (c + as) \right] \\ u = \left[y(s - cx) / (c - as) \right] \\ v = \left[y(s + cx) / (c + as) \right] \\ w = \left[y(s - cx) / (c + as) \right] \end{array} $	$\begin{array}{llllllllllllllllllllllllllllllllllll$
Z - WWW WWWW	CALLICE	IMAGE IMPEDANCES	$k = \left[\frac{y_i + 2z}{v - w}\right]$	Z=z=(tu) ^½ =(vw) ^½	u= v= dz t= w= Dz	Z=z u=v t=w s²=1 d=coth? y=u+w D=tanh? ⊕=0115129 *No.(db)
Will with a		UNIT MATCHED	$k = \left[\frac{\psi' + x_{z}'y_{z}'}{u'v - f'w'}\right]$	$S^{2} = \left[f' \cup \frac{X_{1}' y_{1}'}{X_{1}' y_{2}'} \right]^{\frac{1}{2}}$ $I = \left[f' \cup \frac{X_{2}' y_{2}}{X_{1}' y_{1}} \right]^{\frac{1}{2}}$	t' = [s(s+cx) / (c+as)] u' = [s(s+cx) / (c-as)] v' = [s(s+cx) / (c-as)] w' = [s(s+cx) / (c+as)]	$\begin{array}{llllllllllllllllllllllllllllllllllll$
32 The second se		IMAGE IMPEDANCES	$k = \left[\frac{y_i + 2}{v' - w'}\right]$	l = (†ú') [‡] =(v'w') [‡]	u'= v'= d t'= w'= D	Z = z s2 = 1 t' = t/z x1 = y1 = x2 = y2 = (u'+w') u' = u/z w' = w/z v' = v/z
gtryn junig	BRIDGED T	FULL	$k = \left[1 + \frac{z \left[v(s^{2+1}) + 2s^{2}z\right]}{w(v+2s^{2}z) + s^{4}z^{2}}\right]$	$z = \left[v s^{2} z \frac{s^{2} z + 2w}{v + 2s^{2} z} \right]^{\frac{1}{2}}$	v = 2 zs²/(ds²- 1) w= z (d- s²)/2	$Z \ge z$ $s^2 = Z/z \ge 1$ d = coth $\frac{9}{2}$
		IMAGE IMPEDANCES	$k = \left[1 + \frac{2z(v+z)}{w(v+2z)+z^2} \right]$	$z = \left[v z \frac{z + 2w}{v + 2z} \right]^{\frac{1}{2}}$	v = N z w = n z	$ \begin{array}{cccc} Z = z & s^{2} = 1 & k = E^{\Phi} \\ n = 1/(k-1) = 1/(E^{\Phi}-1) \\ N = (k-1) = (E^{\Phi}-1) \end{array} $
p-min-p	BRIDGED	UNIT MATCHED	$k = \left[1 + \frac{v'(s^2 + 1) + 2s^2}{w'(v' + 2s^2) + s^4} \right]$	$ =\left[v'st\frac{s^2+2w'}{v'+2s^2}\right]^{\frac{1}{2}}$	v'= 2s²/(ds²-1) w'= (d-s²)/2	$Z \ge z \qquad s^{z} = Z/z \ge 1 \qquad z = 1^{\omega}$ $v' = v/z \qquad w' = w/z \qquad d \operatorname{coth}_{\frac{\omega}{2}}^{\frac{\omega}{2}}$
	Т	IMAGE IMPEDANCES	$k = \left[1 + \frac{2(v'+1)}{w(v'+2)+1} \right]$	$I = \left[v' \frac{1 + 2w'}{v' + 2} \right]^{\frac{1}{2}}$	v*= N w′= n	Z = z s ² = J z = Ι ^ω ν'= ν/z w'= w/z
pt yes a start	BRIDGED T	FULL MATCHED IMAGE	$k = \begin{bmatrix} 1 + \frac{v}{Z} \end{bmatrix}$ $k = \begin{bmatrix} 1 + \frac{z}{w} \end{bmatrix}$	$z = (\mathbf{v} \mathbf{w})^{\frac{1}{2}}$	v = N z w = n z	Z = z s²= z²= vw N = (k - i) = E ^e -i = (cosho + sinho - i)
		IMPEDAINCES	$\mathbf{k} = \left[1 + \left(\frac{\mathbf{v}}{\mathbf{w}}\right)^{\frac{1}{2}}\right]$			n = /(k -)= /(E ⁺ I)= /(coshe +sinhe - I)
	BRIDGED T	UNIT MATCHED IMAGE IMPEDANCES	$k = \begin{bmatrix} I + v \end{bmatrix}$ $k = \begin{bmatrix} I + \frac{1}{w} \end{bmatrix}$ $k = \begin{bmatrix} I + (\frac{v}{w})^{\frac{1}{2}} \end{bmatrix}$	l = (v'w') ¹ 2	v'= N w'= n	Z = z $s^{2} = v'w' = $ N = (k-1) = (ξ^{0} -1) n = 1/(k - 1) = 1/(ξ^{0} -1) v' = v/z w' = w/z Θ = 0 15 29 × No. (db)

from expressions 5 and 8, the transmission loss is

$$\frac{d\mathbf{b} = 20 \operatorname{Log}_{10} \mathbf{k}}{= 20 \operatorname{Log}_{10}} \left(s \frac{v' + w' + 1}{w'} \right)$$
(9)

Since,

v' = v/z, w' = w/z and u' = u/z (10) the loss of the network in terms of the full impedance values is from 9 and 10 db=20 Log₁₀k=20 Log₁₀(s(v+w+z)/w) (11)

Solving equations
$$1$$
 and 2 for u' and equating, then simplifying, we obtain

$$(2v'w' + v'^2 - 1)s^2 = w'^2(1 - s^2)$$
(12)

From 9, the factor

$$k = ((v' + w' + 1)s/w')$$
(13)
which solved for w' becomes

$$w' = s(v'+1)/(k-s)$$
 (14)

Substituting this into 12 and solving for v', the unit series impedance adjacent to the unit termination,

$$v' = (k^{2} + 1 - 2ks)/(k^{2} - 1)$$
$$= \frac{k^{2} + 1}{k^{2} - 1} - s \frac{2k}{k^{2} - 1}$$
(15a)

Using 6 in 15,

$$\mathbf{v}' = \frac{\varepsilon^{2\theta} + 1}{\varepsilon^{2\theta} - 1} - s \frac{2}{\varepsilon^{\theta} - \varepsilon^{-\theta}}$$
(16a)

Figures 7 to 12

The power transmission losses and required element values for the *Lattice* and *Bridged-T* networks in symbolical notation, as shown by the headings of the *Tables of Hyperbolic Functions.*

•

$$v^{i} = \coth \theta - s \operatorname{csch} \theta \qquad (17a)$$

$$= \mathbf{c} - \mathbf{a}\mathbf{s} = \mathbf{c} - \mathbf{w}' \tag{18a}$$

where: $k^2 = e^{2\theta}$, $c = \coth \theta$, $a = \operatorname{csch} \theta$, and $s^2 = Z/z$.

The full series impedance is obtained by multiplying each side of equation 15ato 18a inclusive, by z, getting

$$v = z \left(\frac{k^2 + 1}{k^2 - 1} - s \frac{2k}{k^2 - 1} \right)$$
 (15b)

$$= z \left(\frac{\varepsilon^{2\theta} + 1}{\varepsilon^{2\theta} - 1} - s \frac{2}{\varepsilon^{\theta} - \varepsilon^{-\theta}} \right)$$
(16b)

$$= z (\coth \theta - s \operatorname{csch} \theta)$$
 (17b)

$$= z(c-as) = cz - ay = cz - w \quad (18b)$$

where : $y = (Zz)^{\frac{1}{2}}$.

 $w' = s \frac{2k}{k^2 - 1} = s \frac{2}{\varepsilon^{\theta} - \varepsilon^{-\theta}} = s \operatorname{csch} \theta = a$ (19a)

Multiplying each side of equations 19 by z, the full shunt impedance is obtaine as

$$w = zs \frac{2k}{k^2 - 1} = zs \frac{2}{\epsilon^{\theta} - \epsilon^{-\theta}}$$
$$= zs \operatorname{csch} \theta = asz = ay \quad (19b)$$

Substituting 15a and 19a into equation and solving for the unit impedanc adjacent to the termination s^2 ,

$$a' = s^2 \frac{k^2 + 1}{k^2 - 1} - \frac{2k}{k^2 - 1}$$
(20a)

$$=s^{2}\frac{\varepsilon^{2\theta}+1}{\varepsilon^{2\theta}-1}-\frac{2}{\varepsilon^{\theta}-\varepsilon^{-\theta}}$$
 (21a)

$$= s(s \coth \theta - \operatorname{csch} \theta) \qquad (22a)$$

$$= s(cs - a) = cs^2 - w' \qquad (23a)$$

The full series impedance is found b

[Text, continued on page 72; hyperbolic function tables appear on page 57, 58 and 60.]

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-	Coth' 0		/K-1/	254830 754830 188544 83821.8 47157.6	30176.0 20955.4 15397.1 11786.5 9315.32	7544.50 6234.54 5239.85 4464.99 3849.78	2947.65 2947.65 2011.27 2329.15 2090.60	1207.29 1207.29 839.138 616.333 472.277	302.435 210.217 154.598 118.552 93.7910	76.1093 63.0200 53.0532 45.3145 39.1527	34.1974 30.1286 26.7802 23.9550 21.5689	19.5312 17.7787 16.2591 14.9336 13.7697	12.7436 11.8331 11.0220 10.2967 9.64497	9.05699 8.52551 8.04298 7.60375 7.20316	282
5	Csch 20	2 k°	k-1	434.216 217.155 144.760	86.8583 86.8583 62.0386 54.2829 88.2509	43.4254 39.9471 36.1860 33.4024 31.0154	28,1370 28,1370 28,5402 24,1202 22,8504	21.7070 17.3620 14.4649 12.3950 10.8420	8.66671 7.215007 6.177415 5.398111 4.791108	4.3048 3.9062 3.5734 3.2914 3. 0490	2.8384 2.6538 2.4906 2.3450 2.2144	2.0965 1.9897 1.89211 1.80278 1.72063	1.64490 1.57461 1.50937 1.44861 1.39179	1.33860 1.28870 1.24168 1.19963 1.15564	e = 2.7 18
-	Tanh ² –	$\left(\frac{k-1}{k}\right)^{\pm}$	<pre>/ k+1 /</pre>	0.00000 3.31776x10 ⁻⁷ 1.32480x10 ⁻⁶ 2.98253x10 ⁻⁶	5.00051x10 8.28288x10 ⁻⁶ 1.19301x10 ⁻⁵ 1.62409x10 ⁻⁵ 2.12060x10 ⁻⁵ 2.68428×10 ⁻⁵	3.31430×10 ⁻⁵ 3.31430×10 ⁻⁵ 4.00942×10 ⁻⁵ 5.60103×10 ⁻⁵ 5.60103×10 ⁻⁵	7,45632x10 ⁻⁵ 8,48425x10 ⁻⁵ 9,57658x10 ⁻⁶ 1,07350x10 ⁻⁴	1.32549×10-4 2.07130×10-4 2.98184×10-4 4.05821×10-4 5.30113-10-4	2.28001210-4 1.19198,210-4 1.62200,210-3 2.11784,510-3 2.67330,210-3 2.67330,210-3	3.30636×10-3 3.99892×10-3 4.75658×10-3 5.57555×10-3 6.46689×10-3	7.41907x10-3 8.43532x10-3 9.51581x10-3 1.06601x10-2 1.18677x10-2	.013139 .014472 .015868 .017326 .018849	.020428 .022068 .023773 .023541 .025541	.029242 .031181 .033191 .035234 .037341	b), k≥1,
0	Cosh 20	k ² +1	2 k²	0.0000000 1.00000264 1.00001060 1.00002384	1.0004240 1.0006628 1.0009544 1.0001300 1.0001696	1.0002652 1.0003208 1.0003816 1.0003816 1.000480	1.0005964 1.0005964 1.0007660 1.0007660	1.0010610 1.0016580 1.0023862 1.0023502	1.0042400 1.0066350 1.0095594 1.0130210 1.0170136 1.0715738	1.026628 1.032248 1.032248 1.045132	1.060245 1.068660 1.077578 1.097234 1.097234	1.10792 1.11920 1.13108 1.14356	1.17032 1.18464 1.19954 1.23132	1.24818 1.26568 1.28412 1.30280 1.32240	= 6.115129 × No (c
•	Sinh ² 0	$(k^2 - 1)^2$	4 k ²	0.0000000 .00000132 .00000530 .00001192	.00002120 .00003314 .0000650 .0000650 .0000688	.0001326 .0001326 .0001604 .0001908 .0002240			.003175 .003175 .0065105 .0065105 .006508	.013314 .013314 .019211 .022566	.030123 .034330 .038789 .043564 .048564	.05396 .05560 .05554 .07178 .07178	.0022 08516 09232 .09232 .10757	.12409 .13284 .131406 .15140	θ
3	Ι - ε-θ	k - 1	-	0.00000 .001151 .002300 .003448	.004595 .005540 .006884 .009368	.010508 .011447 .012584 .013720 .014855	.01585 .018252 .019382 .019382 .020510	.02160/ .02276 .03395 .03395	.04501 .05594 .07645 .07743 .08799	.10875 .11895 .12904 .13901	.15861 .15861 .17776 .18717	.20567 .21476 .23375 .23264			
1	Tanh ² 0	(k ² - 1) ³	(<u>k²+1</u>)	0.0000000 0.00000132 00000530 0000132	,00002121 ,00003394 ,00006500 ,00006500	.00010739 .00016017 .00019076 .00025973	0/402000, 10023399 00033299 00038285 0000,288	.0004/85 .0005.302 .0008.283 .0011917 .0016225	.0021174 .0033065 .0047570 .006484 .006484	.000020 .013139 .015868 .018849 .02068	.022242 .033191 .037341 .041745 046363	.051200 .051200 .061504 .0661504	.078471 .078471 .084509 .090728 .097118	.110412 .117295 .124332 .131514	
e	Sinh 2 0	K-1	2 k ²	0.00000 .002303 .004605 .004605	.009211 .011513 .013816 .016119 .018422	.020725 .023028 .025331 .027635 .027635	.032242 .034546 .036850 .039154 .041459	.043763 .046068 .069113 .080678	.092234 .11538 .118560 .18525			50260 52851 52851	.58118 .60797 .63509 .66253 .69322 71849	.74704 .77599 .80537 .83511	
No (db) 1		20 log10 eð	20 log 10 k	8588	ත් තිරින්	8 9191	4 2928	ଟ <i>ଖ</i> ମ୍ପଟ୍ଟମ	6 88 88	8 89 8 8 1 8 9 1 1 1 1 1 8 1 1 1 1 1 1 1 1 1 1 1 1 1 1	+ 58688 8 88588	80888	4 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8	808888 808888	2

(Continued on page 58)

PRICE TRUE AND A

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	No (db)		20 log 10 for the	3.50 3.60 3.70 3.70 3.70	3.90 5.0 5.0 5.0 5.0	6.0 8.7.5 8.	9.0 9.5 10.5 10.5	11.5 12.6 13.6 13.6	14.0 15.0 15.0 15.0 15.0	10.0 17.0 17.5 18.0 18.0	20.0 20.0 20.0 20.0 20.0	2222 2222 2222 2222 2222 2222 2222 2222 2222	888888 055055	888888 50.50 5 8	TAL A
	+	Coth ²	$\left(\frac{k+1}{k-1}\right)^{3}$	25.3043 23.9550 22.7138 22.7138 21.5689	20.5111 19.5312 15.5737 12.7436 10.6495	9.05698 7.81848 6.83583 6.04365 5.39563 4.85864	4,40952 4,40952 3,70541 3,42691 3,18583	2.97585 2.79191 2.62992 2.48660	2.24556 2.14376 2.05230 1.96981	1.00220 1.82763 1.71018 1.65909 1.65909	1.56942 1.55004 1.49382 1.46047 1.42968	1.42967 1.37493 1.35057 1.32797 1.30700	1.28752 1.28752 1.25254 1.25254 1.23685	1.19577 1.18385 1.17272 1.17272	
	o.	Sech 20	2 k ³ +	744768 733299 710338 710338	.098802 .687287 .630282 .574957 .522197	.4/201 .426375 .383774 .344757 .30214 .276983	.247862 .221616 .198015 .176846	.140883 .125691 .112113 .0999863 .0991587	.0794954 .0708734 .0631832 .0563229	.0447520 .0447520 .0358893 .0316899	.0251745 .0224375 .0199980 .0178236 .0158856	.0141582 .0126186 .0112465 .0160235 .0089335	.0079620 .0070962 .0063245 .0056367	.0044774 .0039905 .0035565 .0031698 .0031698	
	Q	Csch ² θ	$4 k^{2}$	5.83583 5.49933 5.18931 5.18931 4.90381	4.39568 4.39568 3.40952 2.70541 2.18583	1.7711 1.48660 1.24556 1.05230 .89525 .76619	.65909 .56942 .49382 .42968 .37493	.32797 .28752 .25254 .22219	.17272 .15256 .13491 .11936 .11936	.09370 .083094 .073729 .065453 .058131	.051650 .045905 .046812 .036295 .036295	.028721 .025561 .022748 .020247 .018027	.016053 .014291 .012728 .011337	.008996 .008014 .007138 .005360	
	E	<u> - ε-θ</u>	x 1 x - 1	3.015195 2.94719 2.88292 2.82209 2.76448	2.70972 2.47319 2.4319 2.13167 2.13167 2.00476	1.89808 1.80728 1.72920 1.66142 1.60215	1.50368 1.50368 1.46247 1.42559 1.39244	1.36253 1.31545 1.31085 1.28845 1.28845	1.24926 1.23208 1.21629 1.20175 1.18834	1.17595 1.16449 1.15387 1.1402 1.13488	1.12638 1.11847 1.11111 1.104248 1.097846	1.091869 1.086287 1.081069 1.076186 1.071621	1.067345 1.063339 1.059585 1.056065 1.056065	1.049665 1.046757 1.044026 1.041462 1.041462 1.039051	
4	-	Coth ² 6	$\left(\frac{k^2+1}{k^2-1}\right)^2$	6.83583 6.49933 6.18931 5.90381 5.4000	5.39568 5.39568 4.40952 3.70541 3.18583 2.79191	2.48660 2.24556 2.05230 1.89525 1.76619	1.65909 1.56942 1.49382 1.42968 1.37493	1.32797 1.28752 1.25254 1.25254 1.25219 1.19577	1.17272 1.15256 1.13491 1.11936 1.10572	$\begin{array}{c} 1.09370 \\ 1.083094 \\ 1.073729 \\ 1.065453 \\ 1.058131 \end{array}$	1.051650 1.045905 1.046812 1.036295 1.032282	1.028721 1.025561 1.022748 1.022247 1.018027	1.016053 1.014291 1.012728 1.01133 1.010100	1.008996 1.008014 1.007138 1.005360 1.005666	
t	ת	Csch 20	2 K ^s - 1	1.11609 1.07851 1.04285 1.004282 1.004282 1.004282 1.004282	.94617 .81183 .70274 .61212 .53622	.47136 .41560 .35726 .32515 .28826	.25584 .22726 .179682 .159870	.142246 .126696 .112826 .100492 .089514	.079748 .071052 .053309 .056412 .050269	.044797 .039921 .035577 .031706 .028256	.025183 .022443 .020002 .017826 .015888	.014160 .012620 .011247 .010024 .0003338	.0079623 .0070964 .0053246 .0056368 .0056368	.0044775 .0039905 .00315566 .0031698 .0031698	6 = 2.718282
	θ •	Tanh ² –	$\left(\frac{k-1}{k+1}\right)^{2}$.039519 .041745 .044026 .044026 .044026	.051200 .064211 .078471 .093901 .110412	.127902 .146288 .165463 .185335 .205806	.226782 .248168 .269876 .291808 .313890	.336038 .358178 .380239 .402156 .423869	.445323 .466468 .487257 .507672 .527634	.547156 .566192 .584731 .602741 .622033	.637176 .653575 .659428 .684715 .699458	.713650 .727310 .740431 .753028 .753028	.776686 .787773 .798378 .808505 .818205	.827445 .836283 .84704 .852717 .852717	k 1,
0		Cosh 20	2 k ²	1.34252 1.36570 1.36570 1.40758 1.40758	1,45500 1,58660 1,73925 1,914984 2,116131	2,345354 2,605700 2,900622 3,234026 3,610357	4.034583 4.512356 5.05000 5.65466 6.33434	7.09808 7.95592 8.91951 10.001 <i>37</i> 11.21595	12,57934 14.10966 15.82720 17.75476 19.91793	22.34537 25.06035 28.12597 31.55575 33.40442	39.72266 44.56816 50.00500 56.10535 62.95017	70.63044 79.24780 88.91676 99.76560 111.9384	125.5963 140.9209 158.1155 177.4079 199.0549	223,3428 250,5947 281,1717 315,4790 353,9737	11512 × No (db),
•		Sinh ² θ	4 k ²	.17126 .18185 .19269 .20389 .21551	.22750 .29329 .36963 .45750 .55808	.67268 .80284 .95030 1.11702 1.30517	1.51730 1.75620 2.02499 2.32733 2.66716	3.04904 3.47801 3.95975 4.50068 5.10797	5.78967 6.55482 7.41361 8.37738 9.45892	10.6727 12.0347 13.5630 15.2779 17.2021	19.3611 21.7841 24.5025 27.5528 30.9752	34.8153 39.1239 43.9585 49.3827 55.4691	62.2967 69.9604 78.5576 98.2041 99.0274	111.1716 124.7975 140.0857 157.2398 1176.4866	
I	9	L			.36904 .40434 .43766 .46912 .49881	.52685 .55332 .57830 .60189 .62416	.64519 .66503 .68377 .71816 .71816	.73393 .74881 .76286 .77613 .78665	.80047 .81163 .82217 .83212 .84151	.85038 .85875 .86665 .87411 .87115	.887798 .894075 .900000 .905594 .910875	.915861 .920567 .925011 .923056 .933166	936904 940434 948911 949881	.952685 .955332 .957830 .9620189 .962016	
I	TL2 0		([]])	.146288 .153862 .161569 .169382 .177305	.1 85 335 .226782 .269876 .313890 .358178	.402156 .445323 .487257 .527634 .5266192	.602741 .637176 .669428 .669458 .699458	.753028 .776686 .798378 .818205 .836283	.852717 .867634 .893360 .904390	.914330 .923281 .931333 .938568 .945062	.950886 .956108 .966788 .968726	.972080 .97507 .97757 .980153 .980153	984201 985910 987432 988788 990001	.991084 .992050 .992813 .993680 .994366	
J	Sinh 2 A	07 HUC	2 K ³	.89602 .92717 .95883 .99098 1.02366	1.05689 1.23178 1.42303 1.633146 1.864943	2.121482 2.406174 2.722782 3.075539 3.469103	3.90691 4.400154 4.95000 5.55553 6.25491	7.02729 7.89282 8.86328 9.95125 9.95125	12.53953 14.07418 15.79558 17.72658 19.89281	22.3228 25.04940 28.10819 31.53990 33.39017	59.71007 49.95509 56.0965 62.9423	70.6233 79.2406 88.9112 99.7606 1111.9339	140.9174 140.9174 158.1123 177.4051 199.0524	223.3408 250.5927 281.1699 315.4775 3153.9722	
No (db)		20 log ₁₀ eð	20 log10 k	3.50 3.70 3.80 3.80 3.90	0,4,4,7,7,7,0 7,0,7,0 7,0,7,0	0.7.7.9.99 7.0.7.0.9 7.0.7.0.7	9,0 10.5 11.0 11.0 11.0 11.0 11.0 10.0	11.5 12.0 13.5 13.5	14.0 15.5 15.5 16.0	16.5 17.0 18.0 18.5	210.50 21.0 21.0	33.25 33.25 33.25 33.25 33.25 33.25 33.25 33.25 35 35 35 35 35 35 35 35 35 35 35 35 35	2.5.0.5.0 5.5.0 5.5.0	88888 80505	

(Continued from page 57)

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Gammatron UNIFORMITY MEANS LONGER TUBE LIFE

Heintz and Kaufman engineers have continually developed closer electrical and physical tolerances for Gammatron tubes over the past 16 years, knowing that matched characteristics result in better operation and longer tube life.

Today the importance of tube uniformity, especially in the very high frequencies, is widely recognized; and many of the peacetime standards we have established for Gammatrons are now contained in the wartime specifications for all tubes of the Gammatron type...When you design a transmitter around a pair—or even a dozen—Gammatrons, you will get the full benefit of our years of experience in pioneering constantly higher standards of transmitting tube performance.

HEINTZ AND KAUFMAN LTD. SOUTH SAN FRANCISCO · CALIFORNIA

Gammatron Tubes

HK-254 Matched CHARACTERISTICS

MAXIMUM RATINGS

Power Output	•	×			500 Watts
Plate Dissipation	•				100 Watts
Amplification Facto	r	÷		•	25
DC Plate Voltage	•		×		4000 Volts
DC Plate Current	•	•			225 M. A.
DC Grid Current	a	ž	•	•	40 M. A
Max. Frequency .	•		(n)		175 Mc
INTER-ELECTRODE	CA	P. :			
C grid-plate .					3.6 uuf
C grid-filament		5			3.3 uuf
C plate-filamen	t	•			1.0 uuf
Filament Voltage	•.				. 5 Volts
Filament Current .					7.5 Amps

COMMUNICATIONS FOR JANUARY 1945 • 5

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	(qp) oN	r	20 log₂₀∉ ^θ	20 log10 k	820.0 88.2 810.0 8	33.0 35.0 37.0 37.0	39.0 39.0 41.0 42.0	44.0 44.0 47.0 7.0	48.0 50.0 52.0 52.0	53.0 54.0 55.0 57.0	58.0 59.0 70.0 70.0	75.0 85.0 95.0 95.0 95.0	100.0 105.0 115.0 120.0	125.0 130.0 140.0 150.0	
	+	Coth^{2}	(k+1)	$\left(\frac{1}{k-1}\right)$	1.15256 1.15256 1.14345 1.13491 1.113451 1.110572	1.093697 1.083094 1.073729 1.065453 1.058131	1.051650 1.045905 1.046812 1.036295 1.032282	1.028721 1.025561 1.022748 1.020247 1.018027	1.016053 1.014291 1.012728 1.011339 1.010100	1.008996 1.008014 1.007138 1.005360 1.005666	1.005049 1.004499 1.004008 1.002254 1.001266	1.000713 1.000400 1.000226 1.000126 1.000126	1.000040 1.000022 1.0000128 1.0000128 1.0000074	1.000025 1.0000125 1.0000084 1.0000084 1.0000040	
	r 	Sech 20	2 k³	k ⁵ +	.0022440 .0022440 .00220000 .00158866 .00126191	.00100439 .00079621 .00063245 .00050238 .00039905	.00031698 .00025178 .2x10-4 1.5887x10-4 1.2619x10-4	1.0024x10-4 7.9624x10-8 6.3247x10-8 5.0238x10-8 3.9906x10-8	3.1699×10-5 2.5178×10-5 2x10-5 1.5886×10-5 1.2619×10-5	1.0024x10-5 7.9622x10-6 6.3246x10-6 5.0238x10-6 3.9905x10-6	3.1698x10-6 2.5178x10-6 2.5178x10-6 2x10-6 6.3246x10-7 2x10-7	6.3246x10-8 2x10-8 6.3246x10-9 2x10-9 5.3246x10-10	2x10-10 6.3246x10-11 2x10-11 6.3246x10-11 2x10-11 2x10-11	6.3246x10-19 2x10-19 6.3246x10-14 2x10-14 2x10-14	
	Q	Csch ² θ	4 k ³	(k ² - ,1) ²	.005049 .004499 .004068 .003181 .003181	.002007 .001594 .001266 .001005 .001005	.000634 .000504 .000400 .000318 .000250	.000200 .000160 .000126 .000100 .000082	.000064 .000050 .000040 .000032	.0000180 .0000160 .0000128 .0000096 .0000096	.000064 .000050 .0000050 .0000040 .00000126	.00000014 .00000004 .00000002 .00000002 .00000000		0000000	
	E.	- -		<u>k - 1</u>	1.036787 1.036787 1.034657 1.032655 1.022001 1.025766	1.022900 1.020359 1.018105 1.016104 1.016104	1.012866 1.011347 1.010101 1.008993 1.008007	1.007130 1.006350 1.005655 1.005037 1.005037 1.00487	1.003997 1.003561 1.003172 1.002826 1.002826 1.002518	1.002244 1.001999 1.001781 1.001587 1.001415	1.001261 1.001123 1.001001 1.000563 1.000316	1.0001778 1.0001000 1.0000562 1.0000316 1.0000188	1.0000100 1.0000056 1.0000032 1.0000032 1.0000018	1.0000056 1.0000032 1.0000018 1.0000018 1.00000010	
	2	Coth ² 6	$\left(\frac{k^{2}+1}{k^{2}}\right)^{3}$	$\left(\frac{k^{2}-1}{2}\right)$	1.005049 1.004499 1.004408 1.003181 1.002526	1.002007 1.001594 1.001266 1.001005 1.000059	1.000634 1.000504 1.000400 1.000400 1.000318 1.000350	1.000200 1.000160 1.000126 1.000100 1.00082	1.000064 1.000050 1.000040 1.000032 1.000032	1.000180 1.000160 1.000128 1.000026 1.000096	1.000064 1.000050 1.0000040 1.00000126 1.00000126	1.0000014 1.0000004 1.0000002 1.0000002 1.0000000	1.0000000 1.0000000 1.0000000 1.0000000 1.0000000	1.0000000 1.0000000 1.0000000 1.0000000 1.0000000	292
	57	Csch 20	2 K ³	k ⁴ - 1	.00251786 .00224404 .00220000 .00158873 .00128873	.00100238 .000796214 .000532456 .000502379 .000399052	.000316979 .000251785 .000200000 1.5877x10-4 1.26192x10-4	1.00237x10-4 7.96224x10-6 5.32456x10-8 5.02377x10-5 3.99052x10-5	3.16979x10-5 2.51785x10-5 2.0000x10-5 1.58866x10-5 1.26191x10-5	1.00237x10-5 7.96214x10-6 6.43905x10-6 5.02379x10-6 3.99052x10-6	3.16979x10-6 2.51785x10-6 2x10-6 6.32456x10-7 2x10-7	6.32456x10-9 2x10-9 6.32456x10-9 2x10-9 5.32456x10-10	2x10-10 6.32456x10-11 2x10-11 6.32456x10-13 2x10-13	6.32456x10 ⁻¹⁸ 2x10 ⁻¹³ 6.32456x10 ⁻¹⁴ 2x10 ⁻¹⁴ 2x10 ⁻¹⁴	e = 2.718
	F	Tanh ²	$\left(\frac{k-1}{k}\right)^{3}$	(k+1)	.867634 .874550 .881141 .893360 .904390	.914330 .923281 .931333 .938568 .945062	.950886 .956108 .960788 .960788 .964976	.972080 .97507 .977757 .982293	.984201 .985910 .988788 .988788	.991084 .992050 .992913 .993680 .994366	.994976 .995521 .996008 .997751 .998736	.999288 .999600 .999774 .999774 .999928	096666. 37999966. 32999928. 32999928.	.9999974 .99999874 .99999916 .99999960	k 1,
				-											(a p),
	G	Cosh 20	k°+1	2 k ^e	397.164.28 445.62606 500.00050 629.46300 792.44789	995.63112 1255.9446 1581.1404 1990.5308 2505.9460	3154.7794 3971.6418 5000.000 6.2946x10 ⁸ 7.9244x10 ⁸	9.9762x10° 1.2559x10 ⁴ 1.5811810 ⁴ 1.9905x10 ⁴ 2.5059x10 ⁴	3.1547×10 ⁴ 3.9717×10 ⁴ 4.9999×10 ⁴ 6.2947×10 ⁴ 7.9245×10 ⁴	9.9763×10 ⁴ 1.25593×10 ⁵ 1.58113×10 ⁵ 1.99053×10 ⁵ 2.50593×10 ⁶	3.1547x10 ⁵ 3.97165x10 ⁵ 4.99999x10 ⁵ 1.58114x10 ⁶ 5x10 ⁶	1.58114x10 ⁷ 5x10 ⁷ 1.58114x10 ⁸ 5x10 ⁸ 5x10 ⁸ 1.58114x10 ⁹	5x10 ⁶ 1.58114x10 ¹⁰ 5x10 ¹⁰ 1.58114x10 ¹¹ 5x10 ¹¹	1.58114×10 ¹² 5×10 ¹² 1.58114×10 ¹³ 5×10 ¹³ 5×10 ¹⁴	= 0.115120 × No
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COUPLED CIRCUIT DESIGN

(Continued from page 43)

$$(e')^{a} = e_{p}^{2} + \frac{1}{4}e_{2}^{2} - e_{p}e_{2}S_{2}/U_{2}$$
 (37)

$$(e'')^2 = e_p^2 + \frac{1}{4}e_2^2 + e_p e_2 S_2/U_2$$
 (38)

Assuming the diodes to be operating as peak rectifiers, the rectified voltages are

$$E' = 1.414e'$$
 $E'' = 1.414e''$
 $E = 1.414(e'' - e')$ (39)

For the moment suppose we let $(e')^3 = \alpha - \beta$ and $(e'')^2 = \alpha + \beta$ so that

$$E^{3} = 2[(\alpha + \beta)\frac{1}{2} - (\alpha - \beta)\frac{1}{2}]^{2}$$
$$= [2\alpha - 2(\alpha^{2} - \beta^{2})\frac{1}{2}]^{2}$$
$$E = \pm 2[\alpha - (\alpha^{2} - \beta^{2})\frac{1}{2}\frac{1}{2}$$
(40)

From equations 37 and 38, $\alpha = e_p^2 + \frac{1}{4}e_2^2$ and $\beta = e_p e_2 S_2/U_2$. Using equations 29 and 30 for e_p and e_2 , the d-c output voltage may be written in terms of grid voltage at the driving tube. We are dealing now with scalar values only so the *j* is dropped. First using equation 31 the d-c voltage may be expressed in terms of secondary voltage

losses in the diode, the d-c power generated in the diode load must come from the i-f power supplied by the transformer since no external source of power is available. We can, therefore, equate the power the i-f would produce in a fictitious resistor to the power it does supply in d-c form across the diode load resistor. If we represent the i-f voltage driving the diode by an rms value E, then the peak d-c voltage developed will be 1.414E.

$$E^{2}/R' = 2E^{2}/R = E^{2}/(R/2)$$
 (45)

where R' is the fictitious a-c load across the transformer, which is equal to $\frac{1}{2}R$, *R* being the resistor in the diode load circuit. This applies to a circuit as shown in Figure 14. Another diode rectifier circuit often used is shown in Figure 15. With the same *R-C* time constant the performance will be the same as that of Figure 14 except for the appearance of i-f as well as the d-c voltage across *R*. The resulting power required of the transformer will be

$$E^{3}/R'' = 2E^{2}/R + E^{2}/R = E^{2}/(R/3)$$
(46)

so that, in this case, the equivalent loading on the i-f circuit is (1/3) R.

The discriminator circuit, Figure 9, may be represented by a loaded i-f circuit, Figure 16, in which the effect of the diode loads is incorporated in the fictitious resistors $\frac{1}{2}R$. The total power de-

 $E = \pm 2e_{2} \sqrt{(Z_{o1}/Z_{o2}) (U_{2}/KQ_{g})^{2} + \frac{1}{4} - \sqrt{[(Z_{o1}/Z_{o2}) (U_{2}/KQ_{g})^{2} - \frac{1}{4}]^{2} + (Z_{o1}/Z_{o2}) (1/KQ_{g})^{2}}}$ (41)

A calculation constant is apparent here which is a function of coupling Q, and impedance ratio. Let us call it

$$B = (Z_{o1}/Z_{o2}) (1/KQ_g)^2$$
(42)

From equation 30, at resonance

$$e_{0} = e_{2}/e_{g} = g_{m}Z_{og}[KQ_{g}/(1 + K^{2}Q_{g}^{2})]$$
(43)

and we may now write the discriminator equation in terms of two variables and two constants

$$E = \pm (2A_{o}e_{g}/U) \sqrt{BU_{2}^{2} + \frac{1}{4} - \sqrt{(BU_{2}^{2} - \frac{1}{4})^{2} + B}}$$

where

A

- U_a is the selectivity of a single circuit of O_a
- U is the selectivity of a coupled pair of $Q = Q_{\pi}^2/Q_{\pi}$ A. is the grid-to-grid gain of the loaded
- A. is the grid-to-grid gain of the loaded transformer used as a coupled circuit amplifier (at resonance).
- This reduces discriminator performance to a conveniently calculable formula.

Diode Loads

In the case of a discriminator or any diode-driving transformer, the diode load must be considered in arriving at an effective circuit Q^5 . We will assume that full peak rectification is obtained. Bypass capacitors will, in all usual designs, be sufficient to develop very nearly peak voltage. Where the diode load resistors are low in value, a loss of voltage due to internal diode resistance will need to be anticipated, and this internal diode resistance the loading effect of the resistors. Neglecting the

livered to the diodes is

$$P = (e')^{2}/(R/2) + (e'')^{2}/(R/2)$$
 (47)

Using equation 37 and 38 this becomes

$$P = 4e_{p}^{2}/R + 4(e_{2}/2)^{2}/R$$

= $e_{p}^{2}/(R/4) + e_{2}^{2}/R$ (48)

This means that the power required from the secondary (at any frequency) may be determined from a fictitious resistor of value R across the secondary circuit. The driving voltage for the diodes is

partly provided by the primary, and from equation 48 it is apparent that the power required from the primary at any frequency may be accounted for by a fictitious resistor of value $\frac{1}{4}R$ across the primary. R is the resistance in one diode load.

Let us now apply a sample calculation for Figure 9 with the following circuit constants: Diode load R = 40,000 ohms; Q_1 (coil only) = 90; $C_1 = 40$ mmf; Q_2 (coil only) = 70; $C_2 = 15$ mmf; $f_0 =$ 4.0 mc; $g_m = 3500 \ \mu$ mho; and R_2 (a resistance load across C_2) = 40,000 ohms. The resistance load, R_2 , is not shown in Figure 9.

First we must determine effective circuit Q and impedance.

Resonant impedance: Primary alone $Z_1 = X_1Q_1$ = (1000) (90) = 90,000 ohms Secondary alone $Z_2 = X_2Q_2$ = (2667) (70) = 186,600 ohms

The loading on the secondary is 40,000 ohms in parallel with a fictitious resistor (Continued on page 74)

★ Engineers at Hallicrafters are continually striving for new heights of perfection in high frequency development work. The Model S-37 is one example of the progress they have made. This is the first and only set of its kind - covering both AM and FM and operating in the range of 130 to 210 Mc. Two r.f. stages are used and in conjunction with an intermediate frequency of 18 Mc., assure an amazingly high ratio of image rejection. It is becoming a valuable instrument in the hands of all exploring the upper reaches of the high frequency ranges.

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Personals

NROM Colonel Robert B. Woolverton, veteran of over forty years in the wireless art and inventor of the spark transmitter for which he was made a Fellow of the IRE in 1915, we've received an interesting note describing his early experiences. Col. Woolverton states that his experiments with the Slaby-Arco coherers appeared quite trivial when he began working with electrolytic detectors improvised from miniature lamp bulbs and pieces of Wolleston wire with a pair of headphones, way back in 1904. For the first time, he says, he heard signals! . . . Another real pioneer has also written in . . . Greenleaf W. Pickard. Mr. Pickard began sending out dots and dashes at Blue Hill in 1898-1899, and Providence in 1899. On the schooner Maid of the Mist he reported the 1901 yacht races using the *Pickard-Shoemaker* ten-seconddash code. Frequent unlicensed wireless communication went on between the Pickard station HY at Amesbury, Mass., and operator Charles C. Kolster (now, incidentally, the Inspectorin-Charge for the FCC in Boston) at the Stone Isles of Shoals station S off the coast of New Hampshire. Mr. Pickard received his first official license in 1913.

Commander Fred Muller has written in asking for letters. He says he "gets lonesome at times for home and my old pals." His address is USNR Navy No. 116, c/o Fleet P. O. New York, N. Y. . . . VWOA members H. H. Parker and Ladaveze sent in Christmas greetings. Thanks. HHP continues active up Westchester way with the lighting company and brother Ladaveze continues his technical duties at RCAC at Rocky Point.

Oldtimers

D. GUTHRIE informs us that one of the first men to aid Marconi in his early wireless experiments was, until recently, a resident of New York . . . Captain Arturo Hugo. Captain Hugo served aboard the American Shipping Board vessels in the 1920's. In 1896 he was

Jack Poppele, chief engineer of WOR, who was recently elected president of TBA. He will receive the Marconi Memoral Plaque for TBA at the 20th anniversary dinner-cruise in February.

a Chief Signalman in the Royal Italian Navy serving on the HIMS Carlo Alberta. Upon Marconi's arrival aboard the Alberta to carry on his experiments, Captain Hugo was assigned to assist Marconi, in view of his code knowledge. His work so pleased Marconi that a request to make him a permanent assistant was made. However, Captain Hugo had just passed his examination for a Third Mate's license to enter the Italian Merchant Marine. Later Captain Hugo came to America. Mr. Guthrie recalls meeting him often when he was master of various Shipping Board vessels back in the '20's. . . . William L. Lawson's thirteen-year professional radio record is most interesting. His assignments include service aboard the steamship Fluor Spar, airlines of the Eastern Air Transport, Tropical Radio, Eastern Air Lines, and Pan American Airways. He is now with the Civil Aeronautics Administration.

Los Angeles

A LETTER from Leroy Bremer discloses that he had moved from Alaska down to the Solomon Islands during the year. He is now back in Alaska. Recalling his experiences since Pearl Harbor, he says that he has spent 29 months out of the last 33 outside of the continental United States and covered a halfmillion miles, plus plenty of action on all fronts.

Continuing, he says: "Of course I keep in touch with what's going on, and where the boys are, through the pages of COMMUNICATIONS which always gets to me no matter where I go. Keep up the good work, Bill."

Lieutenant Leroy Bremmer's address is: USMS, Communications Officer, Naneen, Alaska (Bristol Bay).

20th Anniversary

UR twentieth anniversary dinner-cruise in the Belvedere Roof of the Hotel Astor on Saturday evening, Feb. 17, 1945, will feature the presentation of our Marconi Memorial Plaque to the Television Broadcasters Association, of which our own 'Jack' Poppele is the 1945 president. Many additional awards will be made. A full representation of the top ranking personnel of the various Armed services and some of the communications heroes of this war is also planned. . . We'd like to hear from you regarding your reservations at the earliest possible moment.

Biographical Notes

L IEUT. COL. THOMPSON H. MITCHELL who succeeded the late William A. Winterbottom as vice president and general manager of R. C. A. Communications, Inc., has had a varied wireless training.

He was Hawaiian general manage for RCAC from 1930 to 1935, and on the latter date was made manager of the Southern California District offices. He is a graduate of the U. S. Naval Academy at Annapolis (class of '25). He resigned from the Navy in 1927 to enter the communications field. In 1942, he accepted a commission as Major in the Office of the Chief Signal Officer of the Army, and in March, 1943, he was promoted to Lieutenant Colonel.

During his Los Angeles operations Colonel Mitchell was exceedingly active in VWOA activities. Now, he is with our New York chapter. Recently he was awarded the Legion of Merit by the War Department.
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What is the "bottle-neck" in post-war expansion of civil aviation See page 8

Why CAA is installing Ultra High Frequency radio ranges.

See page 8

What anti-collision devices are being developed . . See page 9

What electronic aircraft detectors are See page 9

What can civil aviation learn from the A.A.C.S. . . See page 2

What goes into an instrument landing system . . See page 11

What is approach control.

See page 11



These questions and dozens of others of vital import to all those interested in the development of radio in aviation for increased safety of human life and property are discussed in the pages of

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RESISTIVE NETWORKS

(Continued from page 54)

multiplying each side of these equation by z, obtaining,

$$= z \left(s^{2} \frac{k^{2} + 1}{k^{2} - 1} - \frac{2k}{k^{2} - 1} \right)$$
(20b)

$$= z \left(s^2 \frac{\varepsilon^{2\theta} + 1}{\varepsilon^{2\theta} - 1} - \frac{2}{\varepsilon^{\theta} - \varepsilon^{-\theta}} \right) \qquad (21b)$$

$$= \operatorname{sz} \left(\operatorname{scoth} \Theta - \operatorname{csch} \Theta\right) \qquad (22h)$$

 $= (cZ - y \operatorname{csch} \Theta)$ (23b For the special, but most common case when the terminating impedance are equal, Z = z, or $s^2 = 1$, the unit in z^2 pedances become

$$u' = \frac{k-1}{k+1} = \frac{\varepsilon^{\theta} - 1}{\varepsilon^{\theta} + 1} = \tanh \frac{\theta}{2} = D \quad (24)$$
$$v' = \frac{k-1}{k+1} = \frac{\varepsilon^{\theta} - 1}{\varepsilon^{\theta} + 1} = \tanh \frac{\theta}{2} = D \quad (25)$$

and

u

$$W' = \frac{2k}{k^2 - 1} = \frac{2\epsilon^{\theta}}{\epsilon^{2\theta} - 1} = \operatorname{csch} \theta = a$$
 (20)

Multiplying each side of these equation by z, the full impedance values of the elements are obtained as

$$u = z \frac{k-1}{k+1} = z \frac{\varepsilon^{\theta} - 1}{\varepsilon^{\theta} + 1} = z \tanh \frac{\Theta}{2} = D$$
(27)
$$v = z \frac{k-1}{1-k+1} = z \frac{\varepsilon^{\theta} - 1}{\varepsilon^{\theta} + 1} = z \tanh \frac{\Theta}{2} = D$$

$$w = z \frac{2k}{k^2 - 1} = z \frac{2}{\epsilon^{\theta} - \epsilon^{-\theta}} = z \operatorname{csch} \theta = d$$
(28)

Minimum Loss of the T Network

The minimum loss of this network obtained as the series arm adjacent the smaller terminating impedance as proaches zero. Passing to the limit zero, from 17a or 17b

$$s \operatorname{csch} \Theta = \operatorname{coth} \Theta$$
 (30)

from which

$$s = \cosh \Theta = B$$
 (3)

$$s^3 = \cosh^2 \Theta = E \tag{32}$$

Hence, from either column B or Ethe tables in part II of this series, the minimum loss for a pad having an desired ratio of terminating impedance may be found by inspection. Instead interpolating for any odd values not give explicitly, some engineers may prefer use the relationship between the factor, and the impedance ratios arising from equating equation 150 or 15b, to zero. If use of the quadratic formula, the result

 $k = s + (s^3 - 1)^{\frac{1}{3}}$

By the use of equation 9, 33 gives

 $= 20 \operatorname{Log}_{10}(s + (s^2 - 1)^{\frac{1}{2}}) \quad (34)$

inge Impedances of the T Network

Considering Figure 1, the square root the product of open- and shortcuited impedances taken from the large bedance end is

$$= \left((u+w) \left(u + \frac{vw}{v+w} \right) \right)^{\frac{1}{2}} (35a)$$
$$= \left((uv+uw+vw) \frac{u+w}{v+w} \right)^{\frac{1}{2}} (35b)$$
$$= (\Phi\beta)^{\frac{1}{2}} (36)$$

were:

$$= uv + uw + vw$$
(37)

u + w

 $\frac{u + w}{v + w}$ (5.)

from the small impedance end, is

$$\left((v+w)v + \frac{uw}{u+w} \right)^{\frac{1}{2}} (39a)$$

$$\left((uv+uw+vw)\frac{v+w}{u+w} \right)^{\frac{1}{2}} (39b)$$

$$= \left(\frac{\Phi}{\beta} \right)^{\frac{1}{2}} (40)$$

e image impedances on a unit basis identical in form but with the primes ociated with each element, indicating ision by z the small terminating imnance; the corresponding forms are wn in the chart with Figure 2.

When the terminating impedances are al, 35b and 39b reduce to an identity, d become equal to

$$z = (u(u + 2w))^{\frac{1}{2}}$$
 (41)

bince any generalized network may be olved to an equivalent T, (and as we in part 1 of this series) the image bedances may be found by consideran of the open and short circuited imlances of the network, or

$$= (Z_{o1}Z_{s1})\frac{1}{2}$$
 (42)

$$= (Z_{02}Z_{s2})^{\frac{1}{2}}$$
 (43)

ere the Z's have the meanings given part I.

APPENDIX I

perbolic Functions

The hyperbolic functions of a real varie which are being presented in this ies of articles (with parts I, II and) in tabulated form were obtained m the definitions of the sine and cosine a rectangular hyperbola. These are turn each defined by an infinite series

(Continued on page 82)

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COUPLED CIRCUIT DESIGN

(Continued from page 68)

of 40,000 ohms representing diode load, so that the effective secondary impedance is 20,000 ohms in parallel with the circuit impedance.

$$Z_{02} = (20,000) (186,600) / 206,600$$

= 18.030 ohms

The primary is loaded by a fictitious resistance of 10,000 ohms representing the diode load. It is also effectively shunted by the choke L, Figure 9, the losses of which would have to be accounted for by another parallel resistor if the primary were not so heavily loaded by the diode. With 10,000-ohms load we may neglect the losses in L. The effective primary impedance is therefore

$$Z_{o1} = (10,000) (90,000) / 100,000$$

= 9000 ohms

$$Z_{og} = \sqrt{(9000)(18,030)} = 12,750$$
 ohms.

Effective Q as loaded:

 $Q_{1} = Z_{o1}/X_{1} = 9.0$ $Q_{2} = Z_{o2}/X_{2} = 6.76$ $Q_{g} = \sqrt{(9)}(6.76) = 7.8$ $Q_{a} = 15.76/2 = 7.88$ $Q = Q_{g}^{2}/Q_{a} = 60.8/7.88 = 7.72$

For all practical purposes we could use $Q_{\mathbf{g}}$ in place of Q for this data, but for purposes of illustration that approximation will not be made. A somewhat overcoupled pair of circuits will give the closest approach to a linear discriminator characteristic; in this case we will use 25% overcoupling, C = 1.25. From equation 13 we find

$$1 + C^{2} = (Q_{g}/Q_{n})^{2}(1 + K^{2}Q_{g}^{2})$$
(49)
$$K^{2}Q_{s}^{2} = (1 + C^{2})(Q_{s}/Q_{s})^{2} - 1$$
(50)

 $\begin{array}{c} R^{*}Q_{g}^{*} = (1+C^{*})(Q_{g}/Q_{g})^{2} - 1 \\ = (2.563)(1.02) - 1 = 1.72 \end{array}$

and from equations 42 and 43

$$B = (9000/18,030) (1/1.72) = 0.29$$

A_o = (3500/10^e) (12,750) ($\sqrt{1.72}/2.72$)
= 21.5

If we calculate the discriminator output for a constant input of $e_g = 1.0$ volt, equation 44 will be:

$$E = \pm (43/U) \sqrt{0.29U_2^2 + \frac{1}{4} - \sqrt{(0.29U_2^2 - \frac{1}{4})^2 + 0.29}}$$

U and U_2 are functions of frequency deviation from resonance. From equation 15

$$U = (1/2.563)\sqrt{(2.563 - S^2)^2 + 4S^2}$$

S = (2\Delta f/f_o)Q = [2(7.72)/4000]\Delta f
= 0.00386\Delta f in k

From equation 35

$$U_{2}^{2} = 1 + S_{2}^{2}$$

$$S_{2} = (2\Delta f/f_{o})Q_{2} = [2(6.76)/4000]\Delta f$$

$$= 0.00338\Delta f, \text{ kc}$$

The complete discriminator curve may be calculated from these formulas; however, charts have been prepared to re-

Tabulation of data from the charts and the resultant discriminator voltage. Measured data for the same circuit are also presented.





move much of the labor and obtain rapid results with less chance of error. If we rewrite equation 44

$$E = 2A_{o}e_{g}(D/U)$$
 (51)

$$E = 2A_{t}e_{g}D,$$

$$D = \sqrt{BU_{s}^{2} + \frac{1}{4} - \sqrt{(BU_{s}^{2} - \frac{1}{4})^{2} + \frac{1}{4}}}$$

it will be noted that the function D con-

tains one variable, U_2 , and one parameter, B. This suggests the possibility of producing a universal family of curves, D versus U_2 for specific values of B. This may be carried a step farther by using the variable S_2 instead of U_2 . This has been done, assuming values of S_3 and B in their useful range. The resulting family of curves are plotted in the discriminator chart.

For the example outlined, the solution for chart use will be

$$E = \pm 43 (D/U)$$
 volts, for $e_g = 1.0$ volt

where D is taken from the discriminator chart and U is obtained from a universal

(Continued on page 92)

S			S ₂			Measured	
∆f kc	.00386∆f	U	.00338∆f	D	±E	+ E	- E
50	.193	1.0	.169	.088	3.77	3.3	3.5
100	.386	0.99	.338	.174	7.55	6.8	6.8
150	.579	0.98	.507	.254	11.10	10.1	10.2
200	.772	0.97	.676	.328	14.50	13.4	13.
250	.965	0.99	.845	.396	17.20	16.1	16.2
300	1.158	1.01	1.014	.452	19 20	17.8	17
350	1.351	1.09	1.183	498	19.60	18.2	16 9
400	1.544	1.20	1.352	.535	19.20	17.6	15.0

B

TBA CONFERENCE

(Continued from page 35)

ensely populated areas there is good ason to believe that a solution is in ght for television's severest handicap e cost and difficulty of supplying the satiable appetite of the public for highiality entertainment, stated Mr. Keister.

ELAY SYSTEMS

THE relay and satellite problem was also discussed by Walter S. Lemmon. He said that at the present

mon. He said that at the present age of the art there appear to be only vo methods of transmitting television nages from city to city; the coaxial able which seems to require a very heavy ivestment and development for wide and service, and the chain of relay ations or automatic booster stations rung across the country.

The latter method is not inexpensive, e said, because to furnish a reliable relay /stem these booster stations will have be substantially constructed to withand all kinds of weather, and contain omplex control apparatus. And to be ecoomical they must simultaneously carry ther forms of communication from city o city as *pay load* along with the precious reight of undistorted television images, e pointed out.

Therefore, said Mr. Lemmon, it is my pinion that the economical growth of elay television systems will depend on ne provision of some other form of ay lead for some time to come, at least ntil television is fullgrown.

Describing this pay load activity, Mr. emmon said that IBM have recently reeived construction permits from the FCC o set up an experimental relay system etween New York and Washington and lso New York and Schenectady, and if he initial experimental operation proves atisfactory, an extension of the system cross the country to Chicago and other oints West is planned.

Mr. Lemmon said that inter-city relay ystems for television should be designed accommodate more than one network t a time in order to best serve the ublic interest. These experimental relay ystems are being designed to carry simulaneously at least two high-definition teleision images in each direction, explained fr. Lemmon.

Mr. Lemmon indicated that the present elay system will carry bandwidths of 10 o 12 megacycles, which can be widened o 20 megacycles. F-m transmission will e used along this relay system.

The relay stations will be fully autonatic, said Mr. Lemmon. Over average at terrain these automatic relay stations ill be located about every twenty-eight thirty miles, he said. And over mountinous territory it is probable that the stance between stations may be increased of forty or fifty miles depending upon ne topography of the country.

In multiplexing, it should be possible o carry television programs, several highuality f-m broadcast programs, several accsimile channels for picture transmison and a number of high speed circuits or transmitting the fast impulses of arious business machines, explained Mr. memmon. The IBM Radiotype, developed ist before the war, is already transmitng typewritten messages at speeds of

(Continued on page 76)

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H man Provide

(Continued from page 75)

one-hundred words per minute over ordinary radio circuits disclosed Mr. Lemmon. He expected an increase in this speed with the new methods planned.

TBA Elections

J. R. POPPELE, secretary and chief engineer of WOR was elected TBA president at the annual meeting of the board of directors held at the conclusion of the conference. He succeeds Dr. Allen B. Du Mont. Mr. Poppele was also elected a member of the board of directors.

Robert L. Gibson, assistant to the vice president in charge of advertising and publicity at the General Electric Company, was elected vice president; Will Baltin was reelected secretary-treasurer and O. B. Hanson, vice president of the National Broadcasting Company, was elected assistant secretary-treasurer. Mr. Gibson succeeds Lewis Allen Weiss as vice president.

Three members of the TBA board, whose terms of office expired, were reelected. They are Dr. Du Mont, Curtis W. Mason of Earle C. Anthony, Inc., and F. J. Bingley of the Philco Radio and Television Corporation.

AERONAUTICAL COMMUNICATIONS

(Continued from page 48)

bands behaving in essentially quasi-optical manner, it became possible to duplicate them at relatively short geographical intervals with little or no possibility of serious interference. In addition, as technical skill advanced, the number of channels available in such a band would increase markedly over what at first seemed available. As the ability to increase usable receiver selectivity in terms of improved transmitter frequency stability developed, separation between channels could be decreased, thereby yielding additional usable channels in what at first had to be left as guard bands between adjacent channels.

About 1941 to 1942, fifteen frequencies between 129 and 131.8 mc were assigned, upon an approximately 500-mile duplication basis, to replace the single old 278 kc ground-to-plane communication frequency. Each channel was separated 200 kc. Thus between frequency separation, geographical separation and line-of-sight transmission range limitations the possibility of interference, so prevalent upon the old single 278-kc frequency, were virtually eliminated.

virtually eliminated. Under this plan an aircraft in flight would call and work ground upon a frequency of 140.1 mc. At the start, with a few aircraft in flight, this single frequency would probably serve adequately due to limited aircraft transmitter range, geo-graphical separation and the rule that no aircraft should attempt to communicate with an airport control station until within 10 minutes flight (presumed to be approximately 30 miles) of the airport. With the inescapable tremendous expansion in civilian flying which now seems certain to follow the war, this single frequency will probably have to be expanded to many more, since it is quite conceivable that there may be many airports within some 30-mile areas, not to mention probable need of more than a

(Continued on page 90)



É. F. Johnson Co. Waseca, Mi

POLICE COMMUNICATIONS

(Continued from page 33)

tential police radio licensees in this met-ropolitan region. The probable number of transmitters, fixed and mobile, might approach 10,000. Some of the local police radio engineers claim 500 potential police licensees within a fifty-mile radius of New York city. Additional channels would provide freedom from present interference.

In the Boston metropolitan area similar conditions prevail and similarly in the localities surrounding all large cities in the nation. There are, at present, 27 police radio systems within a 10-mile radius of Boston and 60 stations within a 20-mile radius. Thus it is evident that the actual allocation of frequencies in such an area is a problem for study by competent engineers familiar with the localities in question. In almost every instance these individuals will be found to be the radio engineers for the police departments operating in these areas.

The congested area problem has come about through no fault of the communications engineer. He has had to install police radio systems into an inflexible governmental and police system. The instrumentalities of government, their police, fire, water and sewer systems, their geography and their political and legal organizations are all the product of decades of growth. They are not flexible nor can they be changed to suit the needs of radio engineers. This condition is not defended here but is mentioned because it is a fact so basic in its implications that it is respected and no attempt to evise it is even suggested. The revision of the present governmental set-up is not problem for the radio-communications ngineer.

Wire Facilities Inadequate

Since police radio systems must be enineered to handle peak traffic load; and ince their importance increases with the ncrease in traffic and becomes paramount luring natural disasters, large conflagraions and serious civil disturbances, when vire communication generally fails, it is mperative that some other means of ommunication, other than that depending pon wire lines, be employed. Examples f such situations where ordinary wire acilities were of absolutely no use to he police and emergency organizations re: 1938 hurricane in New England; ocoanut Grove fire in Boston; Detroit ace riots; Port Chicago explosion in California and the recent circus fire in Hartford, Connecticut.

It is a well substantiated fact, suported by evidence in hundreds of cases, hat commercial wire facilities cannot andle emergency police traffic within the me factors involved nor can they supply he area coverage necessary. The cost of wired system producing results at all omparable to a radio-communications ystem is practically prohibitive to the najority of police departments. Furthernore, communication with moving vehiles at great distances is impossible by neans of a wired system and lastly these rired-communication systems are subect to many faults and failures and are articularly vulnerable to weather, accient, sabotage, etc., to which, radio sysem are, to a large extent, immune. Throughout the midwest, south, south-

(Continued on page 78)



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(Continued from page 77)

west and western states, hundreds of police agencies not only make full use of radiotelephone systems for communication to their police cars but must depend, because of the inadequacy, as well as the expense of wired facilities, almost entirely on the radiotelegraph for point-to-point communications. These radiotelephone and radiotelegraph networks are highly developed and handle hundreds of thousands of messages yearly.

Cooperation Between Police

The modern criminal does not confine his activity to one community, to one city, or even to one state. As means of superhigh-speed transportation are developed it becomes more and more necessary that communication between police agencies be as fast, as dependable, and as immediately available, as the communication between the central station and the police cruiser is today.

The criminal recognizes no boundaries and, as a result, the police can operate within no rigid limits. Catching criminals in a radio network woven by instructions and information transmitted to radioequipped police cars, has become a matter of minutes. Time is precious to the lawbreaker—it spells the difference between escape and capture to him. With the use of radio communication between patrol cars and headquarters the criminal is. many times, caught red-handed. Indisputable proof of guilt is frequently obtained, when only circumstantial evidence could otherwise be found.

In cases involving continuous pursuit by a police car in a metropolitan area it is usual to rebroadcast the transmissions from the car. Instructions change so rapidly, as the cars speed past corners and make turns, that no interruption of these communications can be tolerated. These pursuits sometimes last many minutes and happen frequently. Since continuous pursuits in metropolitan areas in many instances lead out into adjacent municipalities, it is well if that town or city can direct their own cars to This requires separate channels. help. Duplex operation is an urgent necessity in large cities for prosecuting fresh pursuit.

Speedy, reliable radio communication between law-enforcement agencies makes possible the direct apprehension of criminals; fast recovery of property; prevention of crime; quick release of innocent persons; prompt aid to the victims of criminal activity and intent; a greater degree of internal efficiency in the police departments and in general, a more effective protection of life and property is afforded the tax payer.

Present Status of Police Radio

At the present there are thirty-one frequencies allocated between 1600 kc and 2490 kc for use by municipal and state police radio stations. Assignment of these frequencies is made within specified geographical areas, to municipal and state police departments.

In the very-high frequencies the FCC rules allocate a total of twenty-nine channels for use by state and municipal police departments. The channels are separated into four groups as follows:

Group A

30700 kc 31900 kc 33940 kc 37500 kc 39900 kc 31100 kc 33100 kc 35500 kc 39100 kc

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Group 35900 kc 37100 kc 31500 kc 33500 kc 37900 kc 39500 kc

Group C

0580 kc 31780 kc 33780 kc 35220 kc 37780 kc 0980 kc 33220 kc 35100 kc 37220 kc 39380 kc

Group D 5780 kc 39180 ke 39780 kc 37 380 kc

Groups A and B are available for ssignment to land and portable municial and state police stations. Portable nd portable-mobile stations may be uthorized to operate on the same freuency as that assigned to the land staon with which their communications re coordinated. Groups C and D are vailable for assignment to municipal and tate police mobile and portable-mobile tations, respectively.

There are at present three medium freuencies and six high frequencies alloated for use by zone and inter-zone olice stations:

RO4 kc	Callina	5195 kc	Calling
808 kc	Working	7480 kc	Dav only
812 kc	Working	7805 kc	Day only
135 kc	Working	7935 kc	Day only
140 kc	Working		

There are also seven frequencies in ne 116-to-119-megacycles band and one requency in the 158-megacycle band, llocated to the police service and availble for assignment on an experimental asis only.

[The new FCC reallocation proposals rovide for 56 channels in the 30 to 44-mc ands. The 152-156-m band was also asigned to this service.]

Up to June 30, 1944, the FCC had censed 1906 municipal-police stations, 52 state-police stations, 88 zone-police tations, 31 interzone-police stations and 2 experimental Class 2 police stations. he number of transmitters licensed, up o the same date, including fixed, portable and portable-mobile stations, were: Aunicipal police, 15,014; state police, 844; zone police, 93; inter-zone police, 2; and experimental, 72.

Generally speaking, police departments hould operate a separate and distinct ommunication system and should be alocated specific channels for their exlusive use. This conclusion is predicated pon several important factors:

(1)—Peak loads occur in every one of he police services, whether in communiation with mobile units or in point-tooint communication. Peak loads must be onsidered as an important factor in alocating frequencies to the police service. Other emergency services, such as fire epartments, transit companies and public tilities, frequently experience peak mesage traffic conditions at the same time s the police departments, as say, a muliple alarm fire in a downtown district r a great natural disaster, hurricane, tc.

When such an emergency arises, all afety and police agencies must act romptly and efficiently. The police deartment must control the traffic on the vay to, and at the scene of, the trouble. f a fire, the fire department must cordinate its forces for the prompt exinguishing of the conflagration, the electric power company must get an mergency crew to the scene to disconnect ive wires and the transit company will robably wish to notify its vehicles of e-routing and schedule changes.

There is always congestion of vehicuar and pedestrian traffic at a fire and

(Continued on page 80)



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(Continued from page 79)

the police must relieve this situation at once or it will seriously hamper the response of the fire apparatus as well as contributing to heavy damage and possible loss of life.

Other conditions where peak message traffic occurs, not only in the police department but in other agencies as well, are large public gatherings outdoors, large parades, serious civil disturbances and natural disasters such as floods and tornadoes. In any of these emergency situations the number of messages required to be transmitted by each agency alone may seriously overload its facilities. Consequently, to attempt to transmit them all over one radio-communication system is obviously inadequate and futile.

(2)—The operating personnel of one service should not be required to give or receive orders from another service. The dissimilarity of operating methods and terminology are such that serious and vital errors are apt to be made. Radio operating personnel familiar with the procedure of their own service and accustomed to executing orders issued in a familiar manner will act as an efficient communication service, which cannot otherwise be achieved.

It should also be recognized that the fact that a police department has a police wire telephone system, if installed in a municipality, has nothing to do with the departments need for a radio-communication system.

Police departments are usually notified of an emergency by public telephone and upon receipt of such messages, they go into action and radio becomes immediately an important part of the police communication system. Mobile units on patrol often call in information of pertinent character with regard to suspiciously acting automobiles or persons and receive instructions at once as to the course of action they are to pursue.

There are some instances of small, isolated rural communities where the maximum peak message load may never reach the point where the radio communication system is overloaded. In these rare cases a coordinated communication system, comprising both police and fire departments may be desirable, both from an operating and an economic point of view. However, if the fire department in such a community is a part of an organized mutual aid¹ network, operating within a fire district, it should then, have its radio communications operating on the fire department frequency allocated to the district or it should cooperate with other small adjacent communities and operate the fire service on a common fire frequency with them.

Economic Factors

The extremely large investment by police agencies all over the country in radio transmitting and receiving equipment would render it highly uneconomica to initiate any large scale changes of frequency allocation.

While the general trend of police radic installation is toward the higher fre quencies there are some localities, which because of geographical situation and peculiarity of terrain to be covered, re quire the use of low frequencies and thei attendant propagation characteristics to

¹ H. A. Friede, COMMUNICATIONS; September 1944. adequately service their area. Such localities are common in California, New Hampshire and other states having mountainous regions within which police radio communication must be maintained.

As present equipment becomes obsolete in the metropolitan areas and in municipalities it is expected that it will be replaced with veryhigh, or ultrahigh-frequency apparatus. Some municipal radio equipment is already operating very satisfactorily in the region near 118 megacycles.

Any proposal to move police and emergency services from their present location to parts of the radio spectrum where the propagation characteristics are radically different must be spaced over a long period of years and should be regarded as a *last ditch* expedient only.

Allocation Needs

Panel 13, of RTPB, requested that police agencies be allocated 31 channels, 8-kc wide, between 1601 and 2490 kc for radiotelephone use. These medium frequencies are, as previously noted, urgently needed in special cases of geographical location and over difficult terrain, where the propagation characteristics of these frequencies especially fit them for solving the communication problems at hand.

the communication problems at hand. In the 30- to 40-megacycle region a total of 86 channels, 40-kc wide, were requested as the minimum number necessary to relieve the terrific congestion in the crowded metropolitan areas. A number of state wide police agencies are now using channels in this region for point-topoint service also.

In the higher portion of the radio spectrum, above 100 megacycles, the panel requested 122 channels, 50-kc wide, between 116 and 156 megacycles. This will permit many municipalities to move from the densely crowded lower frequencies and attain more favorable operating conditions with no loss of efficiency in servicing their areas. The possibility of skipinterference at these frequencies has been found, in over eighteen months of observation to be virtually non-existent

A band of frequencies 10 megacycles wide, from 320 to 330 megacycles, was also requested for miscellaneous police use and experimental communication. It is not expected that this region will become immediately useful but it was thought advisable, in the light of certain restricted information, to reserve some channels in this region for police use.

Certain other channels near 600 mc; 1200 mc; 2400 mc; 4800 mc; and 9600 mc were requested. These requests are new proposals and were not in the RTPB's schedule of frequency assignments. The value of these channels will remain problematical for some time.

The proper perspective for viewing the needs of the law-enforcement agencies of the country for adequate radio-communication facilities can be attained by placing police radio services in their proper category as the nation's *most important* public service organization.

Police departments need adequate communication facilities, not just emergency facilities. This is necessary and essential to the operation of the police departments if they are to render adequate service to the public in police protection. In the final analysis, it is to the adequacy and efficiency of the radio-communication systems of the police departments to which we must look for the promotion of public safety and the protection of property.

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RESISTIVE NETWORKS

(Continued from page 73)

in terms of the mathematical base, ϵ . These series in terms of ϵ are

$$\varepsilon = +\frac{1}{1!} + \frac{1}{2!} + \frac{1}{3!} + \frac{1}{4!} + \dots$$
 (1)

$$e^{\theta} = 1 + \Theta + \frac{\Theta^{a}}{2!} + \frac{\Theta^{a}}{3!} + \frac{\Theta^{a}}{4!} + \dots$$
 (2)

$$e^{-\theta} = 1 - \Theta + \frac{\Theta^2}{2!} - \frac{\Theta^3}{3!} + \frac{\Theta^4}{4!} - \dots$$
 (3)

The value of theta, Θ , as used in this paper is found from the definition of the power ratio which is given in terms of a factor, k. This is

$$k^{3} = W/w = \epsilon^{2\theta}$$
 (4)

The number of decibels gain or loss of any network on a power transmission loss basis is found by taking ten times the logarithm to the common base, 10, of each side of the equalities in equation 4; hence

$$db = 10 \text{ Log}_{10} \text{ k}^{3} = 10 \text{ Log}_{m} (W/w) = 10 \text{ Log}_{10} \varepsilon^{2\theta} (5a)$$

or

$$db = 20 \operatorname{Log}_{10} k = 10 \operatorname{Log}_{m} (W/w)$$
$$= 20 \ \Theta \operatorname{Log}_{10} \varepsilon^{\theta} \quad (5b)$$

Where: W = power input to the network, and w = power delivered to the load.

The common logarithm of the mathematical base, $\epsilon = 2.71828$. . . is 0.43429 . . . , therefore 5b may be written

$$db = 20 \text{ Log}_{10}k = 10 \text{ Log}_{10}(W/w) = 8.685889 \Theta$$
 (5c)

or

$$\Theta = 0.115129 \times \text{No}(\text{db}) \tag{6a}$$

also from 5a $\Theta = \text{Log } \mathbf{k}$ (6b)

$$- \log_{\epsilon} \kappa$$

Calculations of the functions ε^{θ} and $\varepsilon^{-\theta}$ given by equations 2 and 3 for wide ranges of the variable and for small increments of the angle, Θ , were obtained by interpolation from *Tables of The Exponential Function*, ε^{x} , by the Federal WPA for the City of New York, sponsored by the National Bureau of Standards, Washington, D. C.

ards, Washington, D. C. The hyperbolic sine and cosine functions (written, sinh and cosh to distinguish them from the well known trigonometric sine and cosine of circular fuctions which are usually written sin and cos), are defined in terms of equations 2 and 3 as

$$\sinh \Theta = \frac{\varepsilon^{\theta} - \varepsilon^{-\theta}}{2}$$
(7a)

and

$$\cosh \Theta = \frac{\varepsilon^{\theta} + \varepsilon^{-\theta}}{2} \quad \theta \tag{8b}$$

From these two equations, by addition and subtraction, there results $\cosh \theta + \sinh \theta = \varepsilon^{\theta}$ (7b)

- and
- $\cosh \theta \sinh \theta = \epsilon^{-\theta}$ (8b) The sinh and cosh functions of the tables

in part II were calculated from equations

tanh, was obtained from the relation

$$\tanh \Theta = \frac{\varepsilon^{\theta} - \varepsilon^{-\theta}}{\varepsilon^{\theta} + \varepsilon^{-\theta}} = \frac{\sinh \Theta}{\cosh \Theta}$$
(9a)

$$\frac{\cosh \Theta + \sinh \Theta}{\cosh \Theta - \sinh \Theta} = \epsilon^{2\theta}$$

$$= \frac{1 + \tanh \Theta}{1 - \tanh \Theta} = \frac{\coth \Theta + 1}{\coth \Theta - 1} \qquad (9b)$$

The double angle functions were obtained by change of the variable Θ to 2Θ and the definitions as given.

The half angle functions were found from the relationships existing between the sinh and cosh functions of a whole angle, namely,

$$\sinh \frac{\theta}{2} = \sqrt{\frac{\cosh \theta - 1}{2}}$$
(10)
$$\cosh \frac{\theta}{2} = \sqrt{\frac{\cosh \theta + 1}{2}}$$
(11)
$$\tanh \frac{\theta}{2} = \sqrt{\frac{\cosh \theta - 1}{\sinh \theta}}$$
(12a)

or ,

$$\tanh\frac{\theta}{2} = \sqrt{\frac{\sinh\theta}{\cosh\theta + 1}}$$
(12b)

The double angle functions used were:

$$\sinh 2\theta = 2 \sinh \theta \cosh \theta$$
 (13)
 $\cosh 2\theta = \cosh^2 \theta + \sinh^2 \theta$ (14a)
or
 $\cosh 2\theta = 2 \cosh^2 \theta - 1$ (14b)
or

$$\cosh 2 \Theta = 1 + 2 \sinh^2 \Theta \qquad (14c)$$
$$\tanh 2 \Theta = \frac{\sinh 2 \Theta}{(15a)}$$

cosh 2 Θ

or

$$\tanh 2\Theta = \frac{2\tanh\Theta}{1+\tanh^2\Theta}$$
(15b)

The reciprocals of the various functions in equations 7 to 15 were taken, and these are defined as

hyperbolic cosecant,

$$\operatorname{csch} \theta = \frac{1}{\sinh \theta} = \frac{2}{\varepsilon^{\theta} - \varepsilon^{-\theta}} \qquad (16)$$

hyperbolic secant,

sech
$$\theta = \frac{1}{\cosh \theta} = \frac{2}{\epsilon^{\theta} + \epsilon^{-\theta}}$$
 (17)

hyperbolic cotangent,

$$\coth \Theta = \frac{1}{\tanh \Theta} = \frac{\varepsilon^{\theta} + \varepsilon^{-\theta}}{\varepsilon^{\theta} - \varepsilon^{-\theta}}$$
(18)

The half- and double-angle reciprocal (Continued on page 84)

ee Extremes HOUSANDS of these coils have gone to war . . . in the far corners of the earth . . . under the sea and in the sky . . . under the worst conditions.

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(Continued from page 83)

functions of equations 10 to 15 are written in similar manner.

The squared functions were obtained either directly by squaring the value of the function itself, or from derivations of the definitions, from which were found

$$\cosh^2 \Theta - \sinh^2 \Theta = 1 \tag{19}$$

$$\operatorname{sech}^2 \Theta + \tanh^2 \Theta = 1$$
 (20)

and

$$\coth^2 \Theta - \operatorname{csch}^2 \Theta = 1 \tag{21}$$

Through the relationships given in 5a and 5b of

(22)

(23)

$$\mathbf{k}^2 \equiv \mathbf{\epsilon}^{2\theta}$$

and

 $\mathbf{k} \equiv \mathbf{\epsilon}^{\theta}$ in conjunction with equations 5c and 6b, the complete interrelatedness between the network transmission power loss, the parameters or elements of the network, and the Hyperbolic Functions of a Real Variable was established.

A system or symbolical notation was then set up which would further compress the relationships between the parameters of the network and the terminating impedances. This notation makes use of the upper and lower case English letters only without either sub or super-scripts. A choice was made arbitrarily for assignment of basic functions involving the full, double and half angles for both hyperbolic and exponential functions. These were denoted symbolically by upper case or capital English letters; for example $A = \sinh \theta$, $B = \cosh \theta$, $C = \tanh \dot{\Theta}$, etc. Their reciprocals were assigned the same letter in the lower case or small English letters; thus, $a = \operatorname{csch} \theta$, $b = \operatorname{sech} \theta$, $c = \operatorname{coth} \theta$, etc.

Identities between the symbolical, exponential, algebraic and hyperbolic forms are shown by the headings at the top of each column throughout the tables. Thus, with the multiple headings given, all of the transformations from one form to another for any function are given without reference to any text material. For those who prefer to deal with power ratios equal to or less than unity, an additional relationship may be written for each heading by substituting k = 1/r. For example, in the heading involving the hyperbolic sine of the angle, there re-sults the identities, $A = \sinh \theta =$ $(k^2-1)/2k = (1-r^2)/2r = (\epsilon^{2\theta}-1)/2\epsilon^{\theta}.$

APPENDIX II

Gudermannian Functions

Tables of the natural trigonometric functions are usually available from the pages of nearly every electrical, radio or mathematical handbook, whereas tables of the hyperbolic functions are seldom found except in special books or treatises rarely available to the average engineer. Most of the tables on hyperbolic functions have too large an interval for the argument of the angle and are not carried out to enough significant places to be of value for accurate work.

A relationship, well known to mathematicians, but unfamiliar to the average engineer, exists between the numerical values of the circular and the hyperbolic functions. These relationships are im-





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portant, convenient and casily applied in practice.

If the angle or argument assumes values between zero and infinity, the hyperbolic sine, sinh Θ , assumes successive values between zero and plus infinity just as the circular tangent, tan a, does when a is varied from 0 to 90°. The hyperbolic cosine, $\cosh \Theta$, assumes values from one to plus infinity like the circular secant, sec b, and the hyperbolic tangent, tanh Θ , assumes values from zero to unity in the same manner as the circular sine, sin c.

Each of the hyperbolic functions give only singly periodic values, and therefore there must be some single value of the circular angles, a, b, and c which corresponds to a particular value of Θ , such that,

$$\sinh \theta = \tan a$$
 (1)

$$\cosh \theta = \sec b \tag{2}$$

and

 $\tanh \theta = \sin c$ (3)

Squaring each side of equations 1 and 2, and using 19 from Appendix I, we obtain,

$$\sec^3 b - \tan^2 a \equiv 1 \tag{4}$$

remembzering that, sec $b = 1/\cos b$ and tan $a = \sin a/\cos a$, 4 becomes, after clearing of fractions,

$$\cos a = \cos b \tag{b}$$

or

$$\mathbf{a} = \mathbf{b}$$
 (6)

In a like manner, using 2 and 3 with equation 20 from Appendix I, and remembering that $\cos^2 b + \sin^2 b = 1$, after simplifying, we have

$$\sin c = \sin b \tag{7}$$

or

$$c = b$$
 (8)
hence, from 6 and 8,

$$a = b = c = \varphi$$

The angle φ is called the Gudermannian of Θ , and is written

(9)

(10)

$\varphi = g d \Theta$

Therefore, the set of relationships may now be written for the circular and hyperbolic functions which show their inter-relatedness, as

$\sinh \Theta = \tan \operatorname{gd} \Theta = \tan \varphi$	(11)
$\cosh \Theta = \sec \operatorname{gd} \Theta = \sec \varphi$	(12)
$\tanh \Theta = \sin \operatorname{gd} \Theta = \sin \varphi$	(13)
$\operatorname{csch} \Theta = \operatorname{cot} \operatorname{gd} \Theta = \operatorname{cot} \varphi$	(14)
$\operatorname{sech} \theta = \cos \operatorname{gd} \theta = \cos \varphi$	(15)
$\operatorname{coth} \Theta = \operatorname{csc} \operatorname{gd} \Theta = \operatorname{csc} \varphi$	(16)

As an example of the application of these relationships, for 10 db, from table 1, part I, $\theta = 1.15129$; and from table 2, part II, sinh $\theta = 1.42303$. From 11, this equals the tangent of the Gudermannian. From a table of natural or circular trigonometric functions, $\varphi = 54.902^{\circ}$. Using equations 12 to 16 inclusive successively, from the table of circular functions of a 54.902° angle, the hyperbolic functions of the circular angle $\theta = 54.902^{\circ}$ are found to be:

$$\cosh \Theta = \sec \varphi = 1.7392$$

 $tanh \Theta = \sin \varphi = 0.81817$ (Continued on page 86)





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ELECTRICAL INSULATION CO. 84 Purchase St., Boston 10, Mass. (Continued from page 85) $\operatorname{csch} \Theta = \cos \varphi = 0.70275$ $\operatorname{sech} \Theta = \cos \varphi = 0.57497$

 $\cot \theta = \csc \varphi = 1.22223$

and

These values were found by use of Mathematical Tables by H. B. Dwight. These tables are well suited for this purpose as the sin, cos, tan and their reciprocal functions are given directly in hundredths of a degree. This makes it possible to check the hyperbolic functions of the tables or those found elsewhere to an accuracy as great as that of the circular functions, which are usually carried to about five significant figures.

The importance of the Gudermannian function is that given any single hyperbolic function of a real variable; all of the remaining hyperbolic functions may be obtained directly using the relationships shown in equations 11 to 16 from a common trigonometric table of circular functions.

Although, in this series of articles, all of the desired hyperbolic functions are given directly in the tables for relatively small increments of the variable, Θ , there will be instances when the interpolation of some function will give greater accuracy than another one. In these cases, the Gudermannian is particularly useful. By this, it is implied that for any specified loss, if the incremental changes either side of that chosen loss are not approximately equal or linear for any desired function, it would be more advantageous to choose another function entirely which was more linear in the desired range of loss. Then, by use of the Gudermannian Function, the actual desired hyperbolic function can be found to a much greater degree of accuracy.

The inverse operation of finding the natural trigonometric functions from tables of hyperbolic functions can of course be done, but the occasion would seldom arise in network design to use the *Gudermannian Function* in this manner.

APPENDIX III

[Data on L-taper networks, types 2 and 3, are presented in this appendix. Figures referred to appeared in Part II, October, Communications.]

L-Taper Network, Type 2

This is an impedance-transforming network matched by a terminating image impedance at the shunt end only and is shown in Figures 9 and 10, on a full and unit impedance basis, respectively.

Assuming a zero-impedance generator delivering unit current into the unit impedance through a unit external impedance, referring to Figure 10, the power ratio is

$$P_z/P_z = \frac{v_1 + s}{s^4} s^2$$
 (39)

where $w_1 = w/z$ and $s^2 = Z/z$. Since $k^2 = P_z/P_z$,

$$s = {v_1 + s^2 \over s^2} s = s \left(1 + {v \over Z} \right)$$
 (40)

From 40, we may solve for the unit impedance and get

(41)

$$w_1 = v/z = s(k-s)$$



from which, the full value impedance is

(42) v = zs (k - s)

The matched input impedance from Figure 10 is

$$l = \frac{w_1(v_1 + s^2)}{v_1 + w_1 + s^3}$$
(43)

Eliminating v_1 from equation 40 and 43, the unit impedance is found to be

$$w_1 = w/z = ks/(ks-1) = s/(s-r)$$
 (44)

from which

$$\mathbf{v} = \mathbf{z}\mathbf{k}\mathbf{s}/(\mathbf{k}\mathbf{s}-\mathbf{1}) = \sqrt{\mathbf{Z}\mathbf{z}}/(\mathbf{s}-\mathbf{r}) \qquad (45)$$

The transmission loss of this type network is, from 40,

$$db = 20 \log_{10}[s(1 + v/Z)]$$
 (46)

When this network is matched on its shunt end and the terminating impedances are equal, Z = z, or s = 1. Then equations 40, 41, 42, 44, 45 and 46 become,

$\mathbf{k} = 1 + \mathbf{v}/\mathbf{Z}$	(47)
v/z = k - 1 = N	(48)
v = z(k-1) = Nz	(49)
w/z = k/(k-1) = m	(50)
y = z (k-1) = mz	(51)

and the transmission power loss is (52) $db = 20 \text{ Log}_{10} (1 + v/Z)$

Minimum Loss for L-Taper Networks of Type 2

For a match of impedances to exist on the low impedance side and at the same time have minimum loss, this can only occur if the series arm equals zero; hence from equation 42,

$$\mathbf{v} = \mathbf{z}\mathbf{s} \ (\mathbf{k} - \mathbf{s}) = \mathbf{0} \tag{53}$$

from which, we obtain the condition for minimum transmission power loss that

(35)

(56)

$$c = s$$
.

This is the same condition for L-taper networks of Type 1 to have minimum loss. Only in this case, the network has degenerated into a simple shunt-impedance having a value of

$$w = zk/(k-1) = akz/2$$
(54)

where $a = \operatorname{Csch} \theta = \frac{2k}{(k-1)}$.

L-Taper Network, Type 3

This type of network is commonly referred to as an impedance matching pad, and is used to transform impedances on a matched basis so that no reflections take place at either of the junctions of the terminating impedances and the network. It is matched on an impedance basis at both ends by the terminations, and is shown on both a full and a unit impedance basis in Figures 11 and 12, respectively. From Figure 12, we may write for the image impedances,

$$s^{2} = \frac{v_{1} (w_{1} + 1)}{v_{1} + w_{1} + 1}$$
(55)

and

$$1 = \frac{w_1(v_1 + s^2)}{v_1 + w_1 + s^2}$$

(Continued on page 88)



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(Continued from page 87)

where $v_1 \equiv v/z$ and $w_1 \equiv w/z$.

Eliminating w_1 from this pair of equations, and solving the resultant for the series impedance,

$$v_1 \equiv v/z \equiv s\sqrt{(s^2 - 1)} \equiv \sqrt{s^2(s^2 - 1)}$$
 (57)
or
 $v_1 \equiv z_2 \sqrt{(s^2 - 1)} \equiv \sqrt{Z(Z - z)}$ (58)

Eliminating v_1 from equations 55 and 56, and solving the resultant equation for the shunt impedance, we obtain,

$$w_{1} = w/z = s/\sqrt{(s^{2}-1)} = Z/v = Z/\sqrt{Z(Z-z)}$$
or
$$w = zs^{2}/\sqrt{s^{2}(s^{2}-1)} = Zz/v = Zz/\sqrt{Z(Z-z)}$$
(60)

Thus, it may be seen that the element values in this type network are a function of the terminations only.

This network is a special case of a so-called minimum loss *T-network*. When the *T-network* fulfills this condition, it has been degenerated into a *L-taper* network of *type 3*, and is accomplished by allowing the element adjacent to the smaller impedance to be reduced to zero. This element may conveniently be represented by hyperbolic functions, and is

$$u = z (\operatorname{Coth} \theta - \operatorname{s} \operatorname{Csch} \theta) = 0 \qquad (61a)$$
$$= cz - asz = 0 \qquad (61b)$$

when the minimum transmission loss condition is applied. From 61, a most useful relationship is obtained, that is

$s = \cosh \theta = B,$	or	(6lc)
$s^2 = \cosh^2 \theta = E$		(61d)

and

 $\theta = \cosh^{-1} s \tag{61e}$

where

a = Csch
$$\theta$$
 = 2k/(k²-1) = 1/Sinh θ ;
c = Coth θ = (k²-1)/(k²+1);
s³ = Z/z = Cosh² θ = E,

and

 $s = \operatorname{Cosh} \theta = B = (k^2 + 1)/2k.$

Substituting 61d into equations 57, 58, 59 and 60 respectively, we find

he unit series impedance,
$$v_1 = \frac{1}{2} \operatorname{Sinh} 2t$$

$$= \frac{1}{2} G = \sqrt{Z/z} \sinh \theta = As \quad (62)$$

the full series impedance, $\mathbf{v} = \frac{z}{2} \operatorname{Sinh} 2\theta$ = $- \overset{Z}{\mathbf{G}} = \sqrt{Zz} \operatorname{Sinh} \theta = A\sqrt{Zz}$ (63) 2 the unit shunt impedance, $w_1 = 2 \frac{Z}{z} \operatorname{Csch} 2\theta$

> $= 2s^{2}g = \sqrt{Z/z} \operatorname{Csch} \theta = as \quad (64)$ (Continued on page 89)



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NEWS BRIEFS

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W. L. NELSON JOINS ANDREW

Vilbur L. Nelson has become mechanical de-ign engineer of the Andrew Company, Chicago 9. Mr. Nelson was formerly with Western lectric.



BURLINGTON R.R. TO EMPLOY RADIO

The Chicago, Burlington and Quincy Railroad s preparing to install a radio communications ystem to direct switching operations in its Chicago and other large yard terminals, as ioon as the FCC assigns frequencies. U-h-f quipment made by the radio division of Ben-lix Aviation Corporation will be used.

FARADAY ELECTRIC BUYS HOLTZER-CABOT SIGNALS UNIT

CABOT SIGNALS UNIT The Holtzer-Cabot Electric Company, Boston, has sold its signal systems department to the Faraday Electric Corporation, Chicago. Included in the sale, which involves no sale r exchange of capital stock, are the company's patents on signal systems, its signal-systems nventories and movable tools used in the manufacture of this equipment. The Holtzer-Cabot Company will continue to manufacture fractional-horsepower motors, and notor generators. In August, 1942, the com-pany became a wholly-owned subsidiary of Commercial Investment Trust Corporation, of New York.

Vew York.

IRAQ AWAITS TELEVISION SAYS RCA DISTRIBUTOR

Sayid Hafidh of Hafidh Al-Kadi, RCA Victor listributor in Iraq, reported recently that tele-vision is one of the developments of modern cience in which the people of Iraq are most

keenly interested. Mr. Hafidh stated that the Iraq government which provides programs over three stations, one of which is the powerful Baghdad trans-mitter, well-known throughout the Middle East, is keenly interested in television.

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BENDIX RADIO APPOINTS CLAUDE LEACH

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Celanese Celluloid Corporation has char its name to Celanese Plastics Corporation. changed * *

MYKROY BULLETIN

A bulletin, 102, discussing the 191/4" x 291/4" (Continued on page 93)

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Electro Voice DIFFERENTIAL MICROPHONES



RESISTIVE NETWORKS

(Continued from page 88) the full shunt impedance, $w = 2Z \operatorname{Csch} 2 \theta$

 $2Zg = \sqrt{Zz} \operatorname{Csch} \theta = a\sqrt{Zz}$ (65) where $A = \sinh \theta = (k^2 - 1)/2k$; $a = 2k/(k^2-1);$ $G = (k^4 - 1)/2k^2,$ $g = 2k^2/(k^4 - 1)$ and Since $s = \cosh \theta = (k^2 + 1)/2k$ (66)

we may express k in terms of s by solving 66 by means of the quadratic formula and taking the positive sign of the square root since $k > \hat{1}$, and get

$$\mathbf{k} = \mathbf{s} + \sqrt{\mathbf{s}^2 - 1}$$

The minimum attenuation for this net-

work to match both of its terminations is, therefore.

$$db = 20 \text{ Log}_{10} (s + \sqrt{s^2 - 1})$$
 (68)

If the ratio of the terminating impedances is given, the minimum-loss pad that must be used for the network to provide a perfect match with no reflections, can be found by inspection from the column headed E which is symbolical nota-tion for $\cosh^2 \theta$. After locating the ratio in this column, the number of decibels loss may be read from the table either at the right hand or left hand margin. Conversely, if the maximum loss that can be permitted is known, the reverse procedure will give immediately the maximum impedance ratio that may be matched and not exceed the given loss.

(67)





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AERONAUTICAL COMMUNICATIONS

(Continued from page 76)

single *party-line* to control the number of aircraft which will desire to take off from and land at some of the larger airports alone during rush hours.

The allocation of frequencies within this new u-h-f aviation radio band is well depicted in Figure 1. This figure has been abstracted from the CAA Technical Development Report No. 28, May, 1941. The famous checker-board pattern of assignment of radio-range frequencies is shown here. It is based upon dividing the continental United States up into 100-mile squares, with one radio-range station located within each square, then grouping twenty-five such 100-mile squares into one major 500-mile square for purposes of frequency assignment. Upon the seemingly sound theory that each of the twenty-five frequencies which would suffice for the 500-mile square may be duplicated 500 miles away, it became possible to divide the country up into 500-mile major squares, with each such square using identical but discreetly spaced frequencies. A study of this map will reveal that not only has this been accomplished, but that, most adroitly, frequencies individually spaced 200 kc have not been assigned to adjacent squares, but within the 500-mile major square have been juggled so as to give an average separation in frequency between adjacent 100-mile squares the country over of nearly 2500 kc. In no case are adjacent 100-mile-square frequencies closer than 400 kc. This operates effectively to minimize interference, although it is stated that with frequencies repeated only at 500-mile intervals, interference is not anticipated below altitudes of 29,000', which few civilian planes may be expected to reach very soon. Eleven auxiliary frequencies all between 119 and 126 mc, are held in reserve to supplement or replace the twenty-five basic frequencies charted.

So much for a brief glance at the postwar civilian aviation frequency assignment plan as it existed before Pearl Harbor. Since then the frequencies involved have been occupied by war services, and there may be some question as to whether they will ever be returned to their originally intended service after V-Day. Should they not be so returned, some other spot in the u-h-f spectrum will probably have to be found which will accommodate an equivalent frequency assignment plan. Whether such a step will be necessary it is too early to guess. If needed, it today seems reasonable to assume that the assignments will probably be lifted up to somewhere between 300 and 600 mc.

ably be fitted up to somewhere between 300 and 600 mc. [The FCC reallocation proposal provides for the following new assignments: 108-118 mc (air navigation); 118-132 mc (aeronautical mobile service); 420-450 mc (amateurs und air navigation); 450-460 mc (air navigation, temporarily); 1450-1500 mc (air navigation); and 1550-1650 mc (experimental aeronautical mobile services).] It appears inescapable that a practically

It appears inescapable that a practically the certain tremendous increase in civilian aircraft in use in the United States is to be anticipated beginning after V-Day. And the vital importance of radio communications to safety of life in the air is believed to be apparent. There appear to be substantial reasons that an expansion in civilian aircraft will occur. Ignoring completely the popularity of aircraft and flying, significant part air power is playing in the war, complete catching of popular fancy, interest of demobilized military flyers, and strong appeal to

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Interican youth, there seem to be two aramount reasons for anticipating such xpansion, an expansion that will open p on a significant scale, a major new eld for radio-equipment engineering, prouction and sale.

Possibly the first, and most immeiately compelling reason is that aircraft roduction is today America's No. 1 inustry. During the war, this is probably it should be, but after cessation of ostilities a few years hence and with robable rapidly dwindling demand for ilitary aircraft, what does America's Vo. 1 industry then do to fully utilize its roductive plant? What does it do to ontribute to national welfare through ontinuing its employment of the large umber of employable Americans today ngaged in aircraft production? A glance t aircraft figures as of Pearl Harbor, Dec. 7, 1941 is suggestive. As of Dec. 1, 1941, inquiry has revealed that there vere but approximately 486 commercial ir transports and 26,500 civilian itinerant nd other non-military aircraft in the Inited States. Presumed correct, these gures provide food for thought. With overnment borrowing of significant umbers of commercial aircraft to aid the var effort, they suggest that the first ostwar demand upon the aircraft indusry will be for commercial transport and argo planes. Approximately 15,000 transport or like aircraft now utilized by the government may offset the possible proluction which this initial demand could reate. In event of release of these airraft to commercial users in an orderly nanner, their number would go far to atisfy initial demand. Conversion of hese aircraft to be suitable for commerial use could create a new and temporary ousiness in competition with the original producers, who might advantageously levote some of their productive capacity nitially to such conversion.

Such demand alone cannot preserve urrent activity-rate in the aircraft in-lustry after V-Day, even without the hreat represented by 15,000 governmentield transport craft. Heavy as the initial, and continuing commercial demand may pe, in character it is unsuited to support plants specializing in smaller types of aircraft which may find it not feasible o convert rapidly to production of large transport cargo aircraft. It is reasonably certain, in any event, that the demand, domestic and export, for large aircraft cannot keep the entire industry operating at anywhere near its present rate indefinitely in terms of number of aircraft produced and number of workers em-ployed. The search for survival alone by the many plants producing smaller aircraft will force intensive cultivation of the civilian market. The average American is ready for such cultivation in terms of conventional aircraft, seems to anxiously await the practical helicopter, and industry need can find long-time relief only in civilian volume-sale possibilities.

If, for unfavorable reasons of cost and convenience which may be envisaged, the private plane market should prove unable over time to support the expanded industry, the second reason comes to its aid. This takes the form of a combination of heavier, probable annual military aircraft procurement than before the war, and the opportunity inherent in developing government-sponsored flying clubs. So great is the romance of flying that

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BRACH Puratone^{*} ANTENNAS, tested and perfected to meet Army and Navy standards, will again resume their established leadership for Home and Auto Radios, Television, Marine, F.M. and other services.

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American youth will dream of being able to fly for years to come, a dream in most cases not to be satisfied for economic reasons. Uncle Sam may require trained pilots; the pilot material is ready and waiting for training. At some future date, the capacity to produce small trainer planes may even be begging for business. Flying clubs all over the country sponsored by the government appears to offer a possible solution. It supports an industry which it must preserve against future need in purchasing the required planes, preliminarily trains at nominal cost to them the pilots it must also have in reserve, and simultaneously provides outlets for the air-enthusiasm of its youth.

Twice now the allocation of radio frequencies to amateur investigation and employment has stood Uncle Sam in good stead, when as a result the government has found ready and waiting a substantial body of technically skilled radio amateurs in its hours of need—1917-18 and 1941-?. The numerous benefits of government *flying clubs* should be as potentially, maybe even more, valuable.

It is believed that consideration of the foregoing will indicate conclusively the inseparability of aircraft and radio in all the aspects of the former after V-Day. In the face of a steadily increasing number of aircraft going into service, communication (which may satisfactorily be only by radio to an aircraft in flight) becomes even more essential to safety of life in the air—yes, on the ground too than is the motor of an aircraft. Just one glider out of control-contact can be a (Continued on page 92)



Serving the Radio and Electronic Industries with precision engineered products.

Wm.T.WALLACE MFG. CO. General Offices: PERU, INDIANA Cable Assembly Division: ROCHESTER, INDIANA



UNIVERSAL STROBOSCOPE

This handy phonograph turntable speed indicator, complete with instructive folder, is now available gratis to all phonograph and recorder owners through their local dealers and jobbers. As a recorder aid the Universal Stroboscope will assist in maintaining pre-war quality of recording and reproducing equipment in true pitch and tempo. Universal Microphone Co., pioneer manufacturers of microphones and home recording components as well as Professional Recording Studio Equip-ment, takes this means of rendering a service to the owners of phonograph and recording equip-ment. After victory is ours—dealer shelves will again stock the many new Universal recording components you have been waiting for.



• COMMUNICATIONS FOR JANUARY 1945

AERONAUTICAL COMMUNICATIONS (Continued from page 91)

lethal weapon indeed in a crowded air.

The following proposition is therefore respectfully propounded; in civilian aircraft, radio communication equipment is the first and primary requisite. This may appear to be a bold statement, but a little contemplation of possible confusion at a major airport on a Sunday afternoon with but one aircraft bereft of contact with the control tower makes it almost inescapable.

On the assumption that it has been indicated that civilian aircraft expansion after V-Day will be on a major scale, and that the need of radio communication equipment in operative condition is the essential prerequisite to civilian flight, it may be concluded that the radio industry is in the position of being given a new and sizable outlet for its skill,

What the radio and aeronautical in-dustries and the government do with this opportunity is the acute concern of radio engineers, for in this expanded field, life itself is the price of inadequate radio equipment. The primary justification here for radio communication is neither pleasure nor convenience, but the preservation of life itself. It is unthinkable that quality may ever be allowed to fall, engineering or production-wise, into the classification of broadcast receivers. Yet, the makers of broadcast receivers are the seemingly logical producers of aeronautical radio equipment in terms of the probable considerable output which will be required.

There is something of a limit, fortunately, upon the sales expansion possible for aeronautical radio receivers, transmitters, direction finders, positioning

equipment, etc., as the reward for reduced selling price, bringing, as price reduction scems almost invariably to do, deterioration of quality. At least two limits exist: One, little more equipment can be profitably sold than there are aircraft made to use them; second, the equipment must be such quality that it is in operative condition when, properly inevitably, government inspection of aircraft and accessories takes place at frequent periodic intervals.

Only engineering refinement, increased production skill, diminished distribution costs would seem to be the proper devices to be employed to reduce selling price. If these alone be given free reign, the essential of preservation of human life in the air will be better served.

The writer feels that we should have government regulation now, rather than later, after abuses have made it inescapable and punitive in possible effect, to insure maintenance in all airborne radio equipment of a minimum standard of quality and serviceability. It would seem this may be accomplished through modification of the present CAA Approved Type Certificate plan. Under the ATC plan the manufacturer may submit his product to CAA for test, and when and if it is found suitable for air-borne use, an Approved Type Certificate is issued indicating its suitability for use by com-mercial air lines, and civilians, should they be interested.

The writer is convinced that if it may be made mandatory by Federal law that no aircraft, powered or glider, may take off without radio equipment in operative condition, and that such equipment must have previously measured up to rational minimum standards of technical excellence, that a significant service will have been rendered to the American people and the radio industry.

It would seem wise to establish from within the radio industry an aeronautical radio advisory committee to collaborate with the responsible government agency w in the formulation and interpretation of technical standards governing engineering and production best calculated to serve the flying public, and through it the radio industry itself.

COUPLED CIRCUIT DESIGN

(Continued from page 74)

selectivity chart.⁶ Data from the charts and the resultant discriminator voltage are tabulated on page 74, with measured data for the same circuit.

Calculated data are, of course, sym-etrical about fo. The data in the metrical about fo. tabulation are plotted in Figure 17.

References:

¹J. E. Maynard, Tuned Transformers, General Electric Review, Part I, pp. 559-561; October, 1943: Part II, pp. 606-609; November, 1943.

^aTerman, Radio Engineering; p. 93.
^aD. E. Foster and S. W. Seeley, Automatic Tuning, Simplified Circuits and Design Practice, Proc. IRE, vol. 25, pp. 289-313; March, 1937.
⁴J. E. Maynard, U. S. Patents 2,338,527.
^bH. Pader, Theory of the Discrimination.

⁶H. Roder, Theory of the Discrimina-tor Circuit for Automatic Frequency Control, Proc. IRE vol. 26, pp. 590-611; May, 1938.

Paul C. Gardiner and J. E. Maynard, Aids in the Design of Intermediate Frequency Systems; Proc. IRE vol. 32, pp. 674-678; Nov., 1944.

NEWS BRIEFS

(Continued from page 89)

ts of glass-bonded mica ceramic insulation, wn as Mykroy, has been published by Elec-ic Mechanics, Inc., 70 Clifton Boulevard,

on, New Jersey. The new sheets are made in thicknesses rang-from $\frac{1}{2}$ " to $\frac{1}{2}$ ".

PERIMENTAL TELEVISION LICENSE FARNSWORTH

cense to erect an experimental television on has been granted to the Farnsworth vision & Radio Corporation by the FCC. to station will be used solely for experitation and research.

EL-MC CULLOUGH ADVANCES NDERLICH, PIERRI AND BROWN

NDERLICH, PIERRI AND BROWN rge F. Wunderlich has been named general rager of Eitel-McCullough, Inc., of San no, Calif., and Salt Lake City, Utah. uis Pierri, manager of the Salt Lake City It, has become production manager of the re plant at San Bruno. Hewitt V. Wilson, tant manager of the Salt Lake City plant. named manager there. ribert C. Becker, liaison engineer, has re-red to his post as field representative. rin H. Brown, industrial relations and connel director for the past two years, has me head of the sales engineering de-ament.

me head of the sales engineering de-ment. in V. Young, assistant to O. H. Brown, been assigned to the sales engineering de-ment to handle public relations. He will work with Leigh Norton, of the labora-research division, in the publication of a company industrial magazine to appear tly.



G. F. Wunderlich Herbert Becker * *

PHENOL A-N INSERT CHART

A phenol A-N INSERT CHART hart of molded A-N insert arrangements melectrical connectors has been published by wrican Phenolic Corporation, Chicago 50, lois. Serts are grouped by total number of con-in vertical columns, in numerical order, cing from top to bottom and left to right. I standard inserts from one contact to hundred contacts are shown full size. Let or pin arrangements are indicated to-er with wire sizes. Also included are cial cable connections and grounded or the inserts. Mechanical spacing of con-tes and alternative positioning of the in-s with new position numbers are given in a case. Chart is 50" x 38". Copy will be sent on receipt of a request mompany letterhead.

mompany letterhead.

M.C. PRICE LIST DATA

price-list bulletin, 1460, covering twenty-n carbon, dynamic and velocity types in a, stand, throat, lip, hand and cartridge es, has been issued by Universal Microhe Co., Inglewood, Cal.

LE PRODUCTIONS FILES 8 **PLICATIONS**

vision Productions, Inc., a subsidiary of amount Pictures, has filed eight applica-is with the Federal Communications Comn ion for permission to erect a series of ex-mental television relay stations in the east,

mental television relay stations in the east, west and far west. namels 13, 14, 15 and 16 were sought in the Y York City; Buffalo; Peru, Vt.; and De-Mich., areas, and channels 9, 10, 11 and vere requested in the El Paso, Texas; Des nes, Iowa; Chicago, Ill.; and Los Angeles. f., areas.

RY MINTER AN R.C. OF A. SPEAKER ivio distortion was the subject of a paper (Continued on page 94)

SPARE PARTS BOXES

Made as per specification — 42 B 9 (Int) for shipboard use, Electrical and Mechanical. Navy grey finish. Immediate Delivery.

LIST

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	No. 1025-1 12" x 6" x 6"				
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Littelfuse FUSE MOUNTINGS





Single, double and multiple pole mountings.

OPEN TYPE SINGLE POLE MOUNTING No. 351001 (old No. 1060). Black bakelite base. Overall length, 21/8". Shakeproof terminals. One mounting hole.

OPEN TYPE DOUBLE POLE MOUNTING No. 351006 (old No. 1068). Same as above but double pole.

LIGHT WEIGHT SINGLE POLE MOUNTING No. 351003 (old No. 1128). 11/2" x 9/16" x 9/16". Bakelite mounting strip, fibre insulator bottom for metal panel mounting. One mounting hole.

UNIVERSAL FUSE PANEL, NO. 1505 SERIES Standardized units for 10 fuse sizes, any practicable number of poles. Send for blueprints.



COVERED TYPE DOUBLE POLE MOUNTINGS Double Pole No. 351009 (old No. 1237-B). Underwriters' Approved. Fibre-lined, metalshielded cover hinged to bakelite base.

Littelfuse Mountings made for all fuse sizes. Ask for details.

LITTELFUSE INCORPORATED RAVENSWOOD AVENUE, CHICAGO 40, ILLINOIS 200 ONG STREET, EL MONTE, CALIF.



nth degree of stability and endurance necessary to wartime operation.

Think about DX Products for your new receivers and transmitters.



NEM2 RKIFL2

(Continued from page 93)

by Jerry B. Minter, chief engineer of Mea. surements Corporation, Boonton, N. J., presurements Corporation, Boonton, N. J., pre-sented at a recent meeting of the Radio Club America. of

This paper analyzed amplitude versus fre quency characteristics and harmonic distortion. Practical methods of measurement were sug-gested and data on several practical systems were offered.

G. E. BUYS KEN-RAD RADIO TUBE INTERESTS

The General Electric Company has purchased the radio tube manufacturing and plant facili-ties of the Ken-Rad Tube and Lamp Corpora-tion at Owensboro, Kentucky, and at Hunting ton and Rock Port, Indiana, and will take over operation of government-owned plants a Tell City, Indiana, and Bowling Green, Ken tucky.

The sale does not include the Ken-Rad elec tric lamp manufacturing business. Carl J. Hollatz, formerly executive vice presi-dent of Ken-Rad, will manage the new G. I operation, which will be known as the Ken Rad division of the electronics department e G. E. * * G. E.

RMA ISSUES 1944-45 DIRECTORY

The new annual 1944-45 RMA membership an trade directory, with data on personnel, prod ucts, etc., was issued recently. The RM. membership now totals 227.

A. M. WIGGINS NOW WITH ELECTRO-VOICE

A. M. Wiggins has been appointed chief re-search engineer of the Electro-Voice Corpora-tion, South Bend, Indiana. Mr. Wiggins was formerly with RCA Labora-tories in Princeton, New Jersey.



CONCORD FARTS SUPPLEMENT

A 16-page supplement 97, with data on meter, resistors, switches, speakers, relays, volume controls, test accessories, transformers, rhe-stats, etc., has been issued by the Concot Radio Corp., 901 West Jackson Blvd., Chicag 7. Illinois. 7, Illinois.

"E" AWARDS

The United Transformer Company, 150 Varia Street, New York City, and the Hoffma Street,



H. L. Hoffman of Hoffman Radio with the "E" flag.



F. D. Bliley (right) of Bliley Blectric and the three-starred "B" flag.

DX CRYSTAL

GENERAL OFFICES: 1200 N. CLAREMONT AVE, - CHICAGO 22, ILL, U. S. A.

tio Corporation, 3430 South Hill Street, Los seles, California, were recently awarded the ny-Navy "E." second white star for the "E" flag has n won by United Electronics Company, wark, N. J. The Sprague Electric Company, th Adams, Mass., and the Bliley Electric apany, Eric, Pa., received their third white for the "E" flag.

RASCH JOINS SHERRON ECTRONIC

ies V. Barasch has become chief engineer oherron Electronics Company, 1201 Flushing onue, Brooklyn, New York. Ir. Barasch was formerly with Western ctric in charge of the cathode-ray and elec-nic tube equipment design section.

E. PROMOTES PRINCE AND SUITS

id C. Prince, vice president of General atric has been appointed head of the gen-

engineering laboratory. r. C. G. Suits, assistant to the director of research laboratory, has been elected a president and head of the research laborar. C. G. research

r. W. D. Coolidge, vice president and di-or of the research laboratory, since 1940, retired.

AL BEAR NOW RADIART S-M 1 Bear has been appointed distributor sales mager of The Radiart Corporation, Cleveland,

Ir. Bear has been with Radiart for seven rs. Recently he has been acting as operains manager.



B. SINCLAIR IN NEW G.R. POST **B. SINGLAIK IN NEW G.K. POSI** Donald B. Sinclair is now assistant chief incer in charge of circuit development at neral Radio Company, 275 Massachusetts enue, Cambridge 39, Mass. Prior to joining the General Radio engineer-staff in 1936, Dr. Sinclair was a research istant at MIT from 1932 to 1935, and a re-rch associate in 1935 and 1936.



RESS WIRELESS SCHOOL TRAINING AVY STUDENTS

e Press Wireless-Signal Corps radio school (Continued on page 97)

UNDERLICH RUNS 118-MC TESTS



orman B. Wunderlich, of the Motorola com-unications division, conducting 118-me mobile lats in Chicago recently. Mr. Wunderlich re-prted that 118-me was better than 30 to 40 mc.

Only "AIR WOUND" Coils Give You All These Advantages

LESS WEIGHT ... No conventional winding form required-less critical material used in manufacture.

LOW DIELECTRIC LOSS . . . Design incorporates an absolute minimum of extraneous material in winding field.

ADAPTABLE TO ANY MOUNTING ... Ideal for plug-in or other services where mounting problems are involved.

LESS SUBJECT TO DAMAGE ... Nothing much to break. Can easily be repaired without tools, even if bent completely out of shape. Bumper rings or other protective features available for extreme services.

GREATER DESIGN ADAPTABILITY ... Can be equipped with fixed or variable internal or external coupling links, special indented turns for easy tapping, and many other special features.

MORE ACCURATE... Can be wound to more uniform pitch,

Easier to tap at the exact desired point. No coil form to cause dielectric loss.

WIDE RANGE . . . Sizes and types for any application. 10 watts to 10 KW.

Samples to your specifications. Write for details.



Exclusive Export Representatives: Lindeteves, Inc., 10 Rockefeller Plaza, New York, N.Y., U.S.A.

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U.H.F. STANDARD SIGNAL GENERATOR MODEL 84

SPECIFICATIONS

aboratory

tandards

CARRIER FREQUENCY: 300 to 1000 megacycles. OUTPUT VOLTAGE: 0.1 to 100,000 microvolts. OUTPUT IMPEDANCE: 50 ohms.

MODULATION: SINEWAVE: 0-30%, 400, 1000 or 2500 cycles. PULSE: Repetition-60 to 100,000 cycles. Width-

1 to 50 microseconds. Delay—0 to 50 microseconds. Sync. input—amplifier and control. Sync. output—either polarity. DIMENSIONS: Width 26", Height 12", Depth 10".

WEIGHT: 125 pounds including external line voltage regulator.

MEASUREMENTS CORPORATION BOONTON • NEW JERSEY



Featuring new materials and designs, the new Nemco Crystal Holders have easily passed every test to which they have been subjected by the Signal Corps. and crystal manufacturers.

Nemco Holders are designed to prevent deterioration of the crystal by repelling water vapor under tropical conditions.

Because we specialize in the manufacture of Crystal Holders exclusively, we can give you the quality and service to help speed your production.

Write for samples and prices; also request quotations on your requirements for imprinting holders with metallicink.

New NEMCO N5X in No. 6105 and 592 may be obtained in all types of FT-243 Holders.

DAN

U-H-F POWER SUPPLY BY IREDELL EACHUS, JR.

Formerly ESMT Communications Instructor, Moore School of Electrical Engineering, University of Pennsylvania; now on leave.

This paper is based on work carried on by the author at the Moore School

PART II

[In the first installment (December, 1944, COMMUNICATIONS), Mr. Eachus described the design and construction of a power supply for a general purpose r-f source operating on approximately 3000 mc. In this, the concluding installment, Mr. Eachus describes some of the characteristics of the klystron, which is used in the u-h-f r-f system.]

T HE klystron oscillator is similar to most other electronic apparatus in that its ruggedness is electrical, notmechanical. The most easily disturbed part of a klystron is the resonator which is made pliable to facilitate adjustment. It is to this that the concentric lines are attached and by which the tube itself is supported. It is therefore necessary to exercise continuous care in avoiding any strain on the tuner by pulling on the coaxial connectors or by jarring the tube. Short flexible leads fastened to rigid connectors must be used to make the external r-f connections to the tube. Strains on the tuner will not only cause the tube to go out of oscillation but will, if repeated too often, break one of the several glass-to-metal seals, rendering the tube completely useless.

Electrical damage to the klystron may be prevented if but one rule is observed; at no time should the accelerating grid current exceed its rated value. The Sperry S-410 klystron has an accelerating grid rated at a maximum current of 15 milliamperes.

There are several values of accelerating voltage and grid current at which the klystron will operate. A typical group of settings, chosen for maximum power output shows that values of accelerating grid current from 2 to 3 milliamperes is optimum.

Much the same as a gas engine will stall when fed too rich a mixture, the klystron will fail to operate on too high a beam current; therefore, when bringing a klystron into operation low values of grid current should be tried for the best results. The klystron may often be made to operate with the control grid negative although, as the output is reduced, it is not usually operated in this condition.

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NEWS BRIEFS

(Continued from page 95)

icksville, Long Island, which has given bution to Army men during the past two one-half years, has begun courses for is of U. S. Navy scamen. school will hereafter be known as the Wireless Institute, a division of Press less, Inc.

less, Inc. Army c emphasis Army curriculum will be followed, spe-Nemphasis being placed on high power mitter work, antenna technique, "trouble ing" and other practical courses directed "d utility and service in the field.

AC TUBE PACKAGING METHOD

thod of packaging large electronic power for overseas shipments has been de-led by Eitel-McCullough, Inc., at its San ro, Calif., plant. It is said to have been wed by the Signal Corps and the Navy e specification standard for all shipments, natic and foreign, of certain tube types. Fe original package was designed for ship-whe Eimac 450-T.

the Estimate 430-1. E package consists of a set of interlocked iess-steel springs suspended in a gal-red steel X-frame which fits diagonally corrugated cardboard box.



X MARKELL JOINS RCA

E. Markell, for the past four years chief fe vacuum tube section of the U. S. Signal as at Camp Evans, has joined the RCA and equipment division as a specialist on istrial tube applications.
Markell will work under the direction of Thees, manager of the equipment tube con at Harrison, New Jersey.

C BECOMES AIREON MANUFAC-RING CORPORATION

h Aircraft Accessories Corporation will cafter be known as the Aircon Manufacwill ng Corporation.

ne Franke has been appointed director of rtising and public relations. His head-uters for the present will be in Kansas d, Kansas.



INERAL ELECTRONICS CATALOG

A atalog, 101, covering d-c voltmeters, am-ners, milliammeters, microammeters and do-frequency ammeters of the a-c thermo-ole type, has been published by the General a tronics Manufacturing Company, 6014 West whington Boulevard, Culver City, California. .

INDIX-MARINE LAB SHIP

10-foot ocean-going marine laboratory to aid the development of instruments and controls inder construction by Bendix-Marine. mong the equipment to be tested in this

Modern Components for the Modern Radio by Micarta Fabricators Call in a Micarta engineer to consult with you on your postwar planning. Our complete facilities are at your service for the manufacture of Terminal Strips, Coil Forms, Jack Plugs, Battery Plug Assemblies, Resistor Boards, Dialites, and other modern components. Write for our new Catalog No. 101. Serving the Radio and Electronic Industries for over 20 years

icarta Fabricators, Inc

5324 RAVENSWOOD AVENUE, CHICAGO 40, ILLINOIS, TELEPHONE LONGBEACH 9700

lab will be radio direction finders, hydraulic equipment, weather forecasting equipment, etc. The 60-foot floating laboratory will have a beam of 16', a 7' 6" draft, and weigh 59 tons. Its cruising speed will be 10 knots and cruising range, 1,600 miles.

M.I.T. TO OFFER ELECTRONICS COURSES

Plans for a professional course in electronics at M. I. T., emphasizing applications to tele-vision and highly developed production methods in which Philco will cooperate, were announced recently by Dr. Karl T. Compton, M. I. T. president.

Under the new cooperative course, which fol-lows a plan established at the Institute many years ago, leading to the master of science de-gree, students in the department of electrical engineering will spend alternate terms at the Institute and at Philco. Plant experience is

(Continued on page 98)

THERMISTORS

(Courtesy Western Electric) Drving metallic oxide beads for thermistors. COMMUNICATIONS FOR JANUARY 1945 • 97

The inherent stamina of Cinaudagraph Speakers is due to experience in design and manufacturing plus highest inspection standards. In all types of Cinaudagraph Speakers, from small watch-like Handic-Talkie units' to large auditorium speakers, you'll find the same precision, the same painstaking workmanship and the same long-lived faithful reproduction.

Watch Cinaudagraph Speakers after Victory!





We're still up to our ears in critical 'war work but when the war's won we will again be ready

. . To DESIGN, DEVELOP and MANUFACTURE . .

Radio Receivers and Transmitters Industrial Electronic Equipment **Airport Radio Control Equipment** Marine Radio Telephone Equipment

Your inquiries will receive immediate action



Wanted ENGINEERS

- Radio
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Work in connection with the manufacture of a wide variety of new and advanced types of communications equipment and special electronic products.

> Apply (or write), giving full qualifications, to:

R.L.D., EMPLOYMENT DEPT., Western Electric Co. 100 CENTRAL AV., KEARNY, N. J. *Also: C. A. L. Haverhill, Mass. Locust St .. Applicants must comply with WMC regulations

NEWS BRIEFS

(Continued from page 97)

thoroughly integrated with professional the retical instruction. The new course is established as an option

The new course is established as an option the department of electrical engineering, which has conducted parallel courses for more the 25 years in conjunction with the General Ele-tric Company, the American Telephone an Telegraph Company, the Boston Edison Com-pany, the Boston Elevated Railway, and the General Radio Company.

KEN BURCAW BECOMES S-M AT C-D

K. C. Burcaw has been appointed sales man ager of the jobber division, Cornell-Dubilis Electric Corporation. ager of the jobber division, comen-buon Electric Corporation. Mr. Burcaw was formerly sales manager Radiart.



MONACK NOW MYCALEX V-P

A. J. Monack has been elected vice president in charge of engineering, of the Mycalex Cor-poration of America. Mr. Monack has been chief engineer since February of 1942. He has also served with the Radio Corporation of America and the Western Electric Company.



ED CONTENT DEVELOPS ACOUSTICAL METHOD FOR WOR

new method of compensation for acoustical A new method of compensation for acoustical defects in large auditoriums has been developed by Edward J. Content, assistant chief engineer of WOR. Mr. Content's method provides for the re-juvenation of overtones.

R. M. WISE IN NEW SYLVANIA POST R. M. WISE IN NEW SILVANIA FOSI Roger M. Wise has been appointed to a newly created post of vice president in charge of en-gineering of Sylvania Electric Products, Inc. Sylvania's director of engineering for the past two years, Mr. Wise previously served as the company's chief radio engineer for ten years.



WHITE NOW JENSEN RADIO PRESIDENT

Thomas A. White has been elected president and general manager of Jensen Radio Manu-facturing Company. He succeeds W. E. Maxfacturing Company. He succeeds W. E. Max-son, who has retired. Mr. Maxson will remain on the board of

directors.

HAVENS BECOMES S-M OF NOMA CONDENSER DIV.

B. H. Havens has been appointed sales man-ager of the condenser division of Noma Electric Corporation, 55 W. 13 Street, N. Y. 11, N. Y. Mr. Havens will direct merchandising of Noma condensers through sales representatives and parts jobbers in key cities throughout the Luited States United States.



D COMPARISON BRIDGE

COMPARISON BRIDGE mparison bridge for measurement of re-s, capacitors and inductors by compari-with a standard has been developed by Gurevics of Freed Transformer Com-72 Spring St., New York, N. Y. An acy of .5% to 20% is said to be available. trument consists of an a-c bridge, phase-oscillator and vacuum-tube voltmeter. A ndicator in conjunction with a calibrated ndicates percentage of difference between unknown and the standard. The compari-pridge is supplied with 3 frequencies: 60 s, 1,000 cycles, 10,000 cycles. Resistors, tors and condensers being compared to a ar standard range are: capacitors from 25 to 20 mfd; inductors 5 microheuries to henries; resistors from 10 ohms to 5 me-ns. hs.

unit measures 10"x71/2"x81/4". For a-c, 25 volts, 50-60 cycles.



* * *

PACITRON ELECTROLYTICS

Sctrolytic capacitors are scheduled for pro-tion by The Capacitron Company, 849 North Uzie Avenue, Chicago 51, Illinois.

RUTHERS-DUNN RELAYS

RUTHERS-DUNN RELAYS w snap-action relays and vacuum switching ing relays have been announced by Struth-Dunn, Inc., 1321 Arch St., Philadelphia 7. he snap-action relay, 79 XAX, employs an nature that is said to almost complete its vel in either direction before the contacts p into the new position. This feature is said permit its use for vacuum-tube circuits, reurrent protection, pulsing circuits, etc. The standard adjustment using 60-ampere is in the coil at approximately .02 watt re-ts in contact pressures of 5 grams with tacts rated 5 amperes, 115 volts a-c: or 0.5 pere, 115 volts d-c, noninductive. Contact ings up to 10 amperes, 115 volts a-c may be ained with 100 or more ampere turns and orresponding increase in power. A sensitiv-of 0.005 watt with 30 ampere turns is also ainable with reduced contact pressures and ings.

ings. The vacuum-switch keying relay, 78CCA100 s seven poles, including one double-throw le which handles high-voltage r-f currents by ans of a vacuum switch. All high-voltage rts are rounded to reduce corona.



79 XAX Relay TR CENTER CONTACT RECTIFIERS enter contact, a patented plate construction in letal-plate rectifiers, which is said to allow (Continued on page 100)

FROM THE HOUSE OF JACKS

and other radio and electronic components!



PL-204 PL-291A PL-291 JK-48 JK-55

America's largest producer of JK-26 jacks. All models built to strict Signal Corps specifications.

Experience for Sale!

Amalgamated Radio, pioneers in the field, maintain experimental and development laboratories for post-war radio and television equipment. Our components are completely engineered in a self-contained factory equipped with tools of our own design. Years of specialized experience assure high quality products at low cost. Inquiries are invited.







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Drake No. 50N NEON Jewel Min. Bayonet Assembly is ideal where a distinct signal is required and observer is directly in front of instrument panel. Its $\frac{1}{2}$ " smooth clear jewel magnifies and intensifies the illumination from the Neon lamp. Red glass jewel can also be supplied. The No. 51N (without jewel) is applicable where 180° visibility is desirable Both units have built-in resistors for NE51 Neon Lamps operating on standard 105 to 125 volt circuits. These rugged units offer BIG savings in power (1/25 watt), long life (3000 hours), wide voltage range, and great reliability.

SOCKET AND JEWEL LIGHT ASSEMBLIES





Neon Lamp Provides 3000 Hour Continuous Operation Warm Glow Visible from All Angles



This Gothard Pilot Light assembly ac-commodates neon lamps which will burn continuously for approximately 3000 hours, as compared with the approximate 500 hour life of ordinary lamps. It oper-ates on 110 volts and consumes only 1/4 watt. The unbreakable lucite protective cap, designed and made for Gothard ex-clusively, provides perfect light dispercap, designed and made for Gothard ex-clusively, provides perfect light disper-sion of the warm neon glow in all direc-tions. Lucite cap unscrews for lamp change. Bakelite socket. Polished and chrome plated jewel holder. 1" mounting hole. Colors: red, green, amber, blue and clear, Ask for complete information on this and wide range of other Gothard Lights. Lights.

othard

MANUFACTURING COMPANY 1335 NORTH NINTH STREET SPRINGFIELD, ILLINOIS Export Division-25 Warren Street, New York 7, N. Y. Cables-Simontrice, New York

THE INDUSTRY OFFERS ...-

(Continued from page 99) (Continued from page 99) coating protection against destructive at-mospheres, is available in units now being manufactured by Selenium Rectifier Products, Federal Telephone and Radio Corporation, Newark, New Jersey. The center-contact construction is available in the same range of sizes and capacities as the standard petal-type contact.



LANGEVIN 101 AMPLIFIERS

LANGEVIN IUI AMPLIFIERS Amplifiers, 101 types, for medium gain, high-power bridging applications, have been de-signed by The Langevin Company, Inc., of 37 West 65th Street, New York. All models are said to deliver 50 watts to a nominal load impedance with less than 3% rms harmonic distortion at 400 cycles. The gain control provides adjustment over a 40 db range and bridging connections. Chassis is 16-gauge welded steel, zinc plated, and bonderized. Weight approximately 45 pounds.



G.R. U-H-F OSCILLATOR

G.R. U-H-F OSCILLATOR A 100 to 500 mc oscillator, type 857-A, has been announced by General Radio Company, 275 Massachusetts Avenue, Cambridge 39, Mass. Maximum output is said to be 1/2 watt or better over the entire frequency range. The frequency determining element in this oscillator is a "butterfly," a tuned-circuit de-veloped by General Radio. In the butterfly, inductance and capacitance are varied simul-taneously with a single control and no elec-trical contact to the moving element is said to be necessary.

to be necessary. The dial is said to read directly in frequency with an accuracy of $\pm 1\%$. A slow-motion drive is provided. The output circuit is inductively coupled to



100 . • **COMMUNICATIONS FOR JANUARY 1945** the oscillator and the output is controlled by varying the coupling. Otput terminal is a coaxial jack.

coaxial jack. Type 857-P1 power supply, supplied with the oscillator, furnishes filament and plate power and operates from a 115- or 230-volt a-t line, 42 to 60 cycles. An electron-ray tube is used to indicate oscillation. Oscillator is $6\frac{1}{6}$ " x $7\frac{1}{6}$ " x $7\frac{1}{4}$ " overall and weighs $6\frac{1}{4}$ pounds. Dimensions of the power supply unit are $5\frac{1}{2}$ " x $6\frac{6}{6}$ " x $7\frac{5}{6}$ ", weight $9\frac{1}{4}$ pounds.

power supply unit weight, 9½ pounds.



WESTINGHOUSE THIN HIPERSIL

Ultra-thin thicknesses of grain-oriented Hip-ersil have been produced by the engineers of American Rolling Mill and Westinghouse. Prewar standard Hipersil laminations were 29 gauge, or about 14-mils thick. This was followed by 7-mil Hipersil, used for 400-cycle aircraft transforming equipment. A 5-mil thick sheet was then developed. Recently 2-mil and 1-mil thick Hipersil for h-f transformers have been developed.



AMALGAMATED RADIO TELEVISION PLUGS AND JACKS

Plugs and jacks are now being produced by the Amalgamated Radio Television Corp., 476 Broadway, New York, N. Y. Jack types are JK-48, JK-26, JK-55; plug types are PL-54, PL-55, PL-204, PL-291, and PL-291A,



ATHIS ATA AVERY MASKING PROCESS

Die-cut masking stickers have been announced by Avery Adhesives, 451-453 East Third Third

ANTENN COUPLING EQUIPMENT

lf it's an antenna problem, Johnson Engineers can give you the answer. Don't waste power. the answer. Don't waste power. Johnson antenna coupling units insure a perfect match and maximum power transfer. Housed in weather-proof cabinets, they provide an inner door with glass window for observing meter, thereby protecting observer from high voltage.

voltage. Other Johnson products include phasing equipment, concentric line, tower lighting chokes, sampling transformers, inductors, condensers, insulators and similar items. Write for more information and prices.



romsn WIRE COMPANY, INC 15 Park Row, New York City, New York Street, Los Angeles 13, California. The masking stickers can be applied without moistening.

RADIO CITY PRODUCTS SIGNAL GENERATOR

SIGNAL GENERATOR A signal generator, model 704, covering from 95 kilocycles to 100 megacycles, has been pro-duced by Radio City Products Company, 127 West 26th Street, New York 1, N. Y. Fun-damental frequencies are said to be continu-ously variable from 95 kilocycles to 25 mega-cycles in 5 bands. Calibration is said to be accurate to 2% per band up to the broadcast band and within 3% for high-frequency bands. Model has a planetary-drive condenser with direct reading calibration. Output can be modulated or unmodulated.



G.E.M. METERS

Meters featuring linear scales and unit con-struction have been announced by General Electronics Manufacturing Company, Culver

Electronics Manufacturing Company, Curver City, California. Unit construction is said to consist of jewel bearings, armature and core assembled as a unit construction that maintains accurate alignment and permits almost instant removal. Alnico magnet and pole piece used are said to be brazed togethet. Meter weighs .5 pound. Standard meters include d-c voltmeters, am-meters, milliammeters, microammeters, and radio-frequency ammeters of the a-c thermo-couple type.

couple type.

THREAD PROTECTORS

Internal and external safeguards for threaded and machined parts, fittings, etc., have been developed by Precision Paper Tube Company, 2033 West Charleston St., Chicago 47, Illinois. They are made of spiral-wound fibre, kraft paper, or cellulose acetate, under heat-treated compression

compression. Protectors are cylindrical, square, rectangu-lar or tapered, crimped, flared or perforated. Tolerances are said to be $\pm .003''$.



I.C.C. CONDENSER MOUNTING Brackets, M type, designed to permit the mounting of oil capacitors in either vertical or inverted position, have been developed by In-dustrial Condenser Corporation, 3243-65 North California Avenue, Chicago 18, Illinois.



G. E. DELTABESTON FLAMENOL LEAD WIRE A new deltabeston flamenol thermoplastic in-(Continued on page 103)





E INDUSTRY OFFERS ...-

(Continued from page 101)

ted lead wire for use in all types of rescent lamp ballasts has been announced by

he insulation of this new wire is said to be eraging and resistant to flame, oils, acids alkalies. The wire is available in solid and inded conductors, sizes 16 and 18 AWG in hy colors, including black, white, red and Pn. * * *

R NARROW LEVER KEY

A NARROW LEVER KEY /16" wide lever key, FTR-810, is now being blued by Federal Telephone and Radio Cor-ation, Newark, New Jersey. and to have 18-spring capacity. Designed one or two way, locking or non-locking ration. Entire key assembly is said to be i together by a single screw. The spring sed steel frame, with all front position ings in one group and all back position matct springs are of nickel silver with postrigs are provided for tension adjust-it. Non-click buffer springs are supplied use in circuits where spring backlash is be avoided. All springs are interchange-and the pile-ups may be rearranged. and the pile-ups may be rearranged.



ELECHRON RADIO PRESELECTOR

semi-automatic radio clock preselector, that prks on the alarm clock principle, for instal-tion in radio receivers, has been developed by Warren Telechron Company (Ashland, ass.).

The telechron alarm clock provides the basic echanism for the new device. It is operated a two-watt telechron a-c synchronous motor d can be furnished for 115-volt or 220-volt eration in any standard commercial frequency.

* * *

VESTINGHOUSE OSCILLOGRAPH

self-contained electronic oscillograph has en developed by Westinghouse Electric and anufacturing Company, P. O. Box 868, Pitts-

anufacturing Company, P. O. Box 868, Pitts-irgh, Pa. Cathodes of the tubes are energized from a -kv d-c rectifier. Unit consists of the oscillograph proper in ort of the cohingt and the cabingt proper

-kv d-c rectifier. Unit consists of the oscillograph proper in ont of the cabinet and the cabinet proper



which houses all energizing and control cir-

in addition to fluorescent screen for direct observation, the instrument contains a sta-tionary film holder taking a standard film for recording 1/1000 of a second or less. May be operated with a rotating film drum for use from 1/1000 to a second. from 1/1000 to 1/10 of a second.



OHMITE ARMY-NAVY AIRCRAFT RHEOSTATS

KHEUSIAIS Two approved power rheostats for aircraft made in accordance with the latest Army-Navy Aeronautical Specifications AN-R-14a, have been announced by Ohmite Manufacturing Co., 4835 Flournoy St., Chicago 44, Ill. The rheostats are light weight. Two sizes are available: model "J" 50-watt and model "H" 25-watt; linear or taper wire-wound, in various resistances, with "off" position, as required.



CLARK LEVER-LOCK BORING BAR

Lever-lock boring bars, available in sets of four . . . 3/16"x5", ¼"x6", ¾"x7" and ½"x8", have been announced by Robert H. Clark Company, 9330 Santa Monica Boulevard, Bev-erly Hills, California.

* * *

Company, 9300 Santa Induct Inductivity, Deterly Hills, California. These bars are said to be designed for use without bushings or adaptors. One precision ground high speed threading bit and one boring bit are provided, with each bar. The lever-lock holds the tool bit at a right angle on one end and at a 15° angle on the opposite end.

INDEX

(Continued from page 66) Vacuum Tube-Cross Reference Indices of New and Old Designations; J. R. SchoenbaumJan. Values of the Function: cos(h sin V)cos V; OsmundsenAug. Α.



***** Hundreds of thousands of these Clarostat power rheostats are now in daily use. Especially so in aircraft assemblies. Indeed, they are standard equipment in planes, radio, electronic and industrial equipment. They are proving that "They can take it"—and then some.



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68



(Continued from page 46)

tween the anode and the cooling mechanism, which is normally at ground potential. Porcelain or rubber tubing is used to insolate the jacket from the rest of the cooling system electrically. Provision must be made also for controlling

electrolysis. This is usually accomplished by providing a metaliic electrolysis target which can be replaced after it has been disintegrated. The necessity for pure water in the system is obvious, since impurities lead to increased conductivity, in-



Variation of current with temperature for tungsten filament for half (B) and full (A) applied voltage. Normal operating temperature equals 2500° K. (Courtesy FTR).



posits which are formed on the anode of the tube.

In forced-air cooled tubes the heat is conducted away from the anode by a copper-fin structure that is soldered to the anode. The heat in the fins is in turn dissipated into the air which is forced through the fin structure. Such a tube is set in an insulated socket which provides a duct for the air flow from the blowers. It is important that the air be clean so that no dust collects in the various parts of the circulating system. Such dust may reduce the efficiency of the cooling apparatus to a dangerous point.

The other seals of the tube which provide for connection through the glass envelope to the filament strands and the grid are made by using a metal with essentially the same expansion coeffi-cient as that of glass. Care must be taken that the temperature of these seals is kept below the danger point. In certain tubes air-blast cooling of the seals may be necessary. Too much cooling, however, may set up strains in the glass, which is a poor conductor of heat.

Although the filament of an externalanode vacuum tube is little more than a strand of tungsten arranged so that it can be heated to inncandescence, its operational characteristics have furnished material for many pages of discussion. The resistance of a tungsten filament at 2500° K is thirteen times as great as the resistance of the same filament at room temperature. When it is realized that many of these filaments draw up to 100 amperes at normal operating temperature, the difficulties in designing a starting



rcuit to bring these tubes to this tem-crature become apparent. The problem two-fold. First, the cold filament is in fect a direct short on the supply; therepre, some means for current limitation lust be supplied. Second, shock to the lament caused by such a high current ansient must be prevented.

Figure 5 shows how the current of tungsten filament varies with temperaare for full and half voltage. It has been nown⁵ that such a set of curves can upply the information necessary to degn the usual filament starting circuit. The two factors to be considered in such esign are first, a means of limiting the urrent surges to twice the rated current; nd second, the total time necessary to ring the filament to operating temperaare. Since there is very little call for apid-start circuits using external anode ibes, the first design requirement is most nportant. It may be seen from Figure 5 hat if the series starting resistance is horted out at point C, the current surge rill be within the requirements. The esistance necessary to give half voltage t this temperature, 1400°, is given by

$$R = \frac{E_F}{2 I_F}$$
(1)

where $E_{\mathbf{F}}$ is the rated filament voltage, and

IF is the rated filament current

bince this resistance is usually inserted n series with the primary of the filament eating transformer, the above relation nust be modified to give the necessary rimary resistance:

(2)

$$=\frac{E_{L^2}}{2E_{L}}$$

where EL is the primary voltage This resistor also limits to a safe value

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the starting current when the filament is cold. The total current through the cold filament must be less than 200% normal, since the resistance determined in 1 is half the operating resistance. High reactance transformers may also be used to limit starting current.

The time necessary for the filament to come to the 1400° temperature with the starting resistance determined from 1 or 2 is difficult to determine because of the number of thermal factors involved.

Jones and Langmuir⁶ have derived an expression for the heating time for tungsten filaments which is accurate where small temperature changes are involved. Experimental determination of the actual heating time for each filament is generally necessary, however. It is interesting to note that, in general, the heating time varies with the square of the strand diameter and with the first power of the length. Since there is no reason for hurrying the process, it is possible to make the first part of the heating cycle long enough to bring the filament to a stable temperature, a condition which can be visually noted by the operator.

To obtain maximum life from a tungsten emitter the operating temperature shuld be kept as low as possible. Figure 6 shows graphically this relationship be-tween temperature and length of life. The extent to which the filament voltage may be reduced is limited by the available emission. In the fundamental design of the tube a compromise is made between the life expectancy on the one hand and filament heating power required on the other. Although tubes with a life of 50,000 hours are within the realms of possibility, they would be very uneconomical because of the increased filament heating capacity necessary, the increased power consumption, and the increased filament heat dissipation.

It is apparent from Figure 6 that the filament life of an external-anode vacuum tube depends not only on the tube design itself, but also on the design of the equipment into which it is installed. This fact is illustrated by the long life obtained from the tubes used in high-level modulated broadcast transmitters where the design has been very generous. Because the external anode tubes are expensive items it is important that the designer consider this life factor in equipment design and balance it against increased costs of conservative operation. [To be continued]



Contraction of the South of the second

CONTROL UNITS ... Now Ready Also an Attractive Binder to Hold Bulletins . On Request

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Figure 6 **Relationship between**

tungsten filament

temperature and length of life.



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for the 5th time

GENERAL RADIO was granted the coveted Army-Navy "E" Production Award for the fifth time in December of last year. G-R is one of the very few in the electronic industry to receive this honor five times.

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G-R is proud that it has been able to meet the urgent production requirements of the war effort. It is equally grateful that the substantial contributions from its Development and Engineering Departments, through many thousands of hours of consulting engineering on secret war projects, have directly assisted in the solution of technical problems of the greatest urgency.

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