

# Balanced Modulators

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THE balanced modulator is a special form of mixer circuit to which a carrier voltage and a modulating voltage are applied simultaneously and in which the carrier is suppressed and only the sideband products appear in the output. The balanced modulator finds application in test instruments, electronic control equipment, and in suppressed-carrier, carrier-current, and speech-scrambling communication circuits. Acquaintance with this type of detector is essential to an understanding of many modern electronic systems.

#### Circuit Configuration

Figure 1 shows the basic arrangement of the balanced modulator. The grids of  $V_1$  and  $V_2$  are 180 degrees out of phase for input  $e_1$ , the modulating voltage, and are in phase for input  $e_2$ , the carrier voltage. The modulating voltage thus sees pushpull grids, and the carrier parallel grids. The grid bias (B<sub>1</sub>) and plate supply voltage (B<sub>2</sub>) are proportioned so that the tubes operate in the square-law region of their dynamic characteristics. Analytically, the circuit is considered to consist of two single-tube square-law modulators in back-to-back connection.

#### **Basic** Operation

The carrier frequency (fc) usually is higher than the modulating frequency (fm). Thus, it often, but not necessarily always, is true that fc is a radio frequency while fm is an audio frequency. The circuit output consists of the upper and lower sidebands, fc-fm and fc-fm. The carrier frequency, fc, is suppressed by the symmetrical circuit and accordingly does not appear at the output terminals.

Performance of the balanced modulator resembles somewhat the characteristics of a pushpull amplifier wherein only odd-numbered harmonics resulting from tube operation appear in the output, although evennumbered harmonics are present in the plate current. Plate current changes, in opposite directions, in the primary of transformer  $T_3$  (Figure 1) produce the output voltage,  $e_3$ .

Application of the low-frequency modulating voltage,  $e_1$ , to the grids out of phase, through the center-tapped secondary of transformer  $T_1$ , causes a half-wave pulse of carrier frequency, fc, from plate to plate, for each half-cycle of  $e_1$ . Symmetrical operation prevents distortion products, due to non-linear operation of  $V_1$  and  $V_2$ , from reaching the output.

Since the carrier,  $(e_2, f_2)$  is applied to the two grids in the same phase, this component is balanced out of the circuit, the completeness of suppression depending upon the symmetry of the circuit.

Sufficient symmetry for complete suppression of the carrier actually is difficult to obtain in the simple circuit of Figure 1 because of the effects of wiring capacitances, dissimilar tube characteristics, and unbalance in the transformers. In order to adjust the circuit for close balance, capacitive and resistive adjustments are included, as shown in Figure 2. Here, potentiometers R<sub>6</sub> and  $R_{12}$  are the resistance balances, and the dual trimmer,  $C_6$ - $C_7$ , is the capacitance balance. With the modulating voltage removed and only the carrier voltage applied, the various balancing controls are adjusted for carrier null at the OUTPUT terminals.

## Typical Tube Circuit

Figure 2 is a pentode-type balanced modulator circuit of the variety em-

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ТҮРЕ	MANUFAC- TURER	Max. Operating Inverse Voltage	Max. Peak Current (ma.)	Max. Average Current (ma.)	Max. Surge Current (ma.)	FORWARD		REVERSE	
						Test Volts	Cur- rent (ma.)	Volts	MA.
1N40	CBSH	25	60	22.5	100	1.5	12.75	10	0.035
1N40	SYL.	25	60	22.5	100	1.5	12.75	10	0.040
1N41	CBSH	25	60	22.5	100	1.5	12.75	_10	0.040
1N41	SYL.	25	60	22.5	100	1.5	12.75	10	0.040
1N42	SYL.	50	60	22.5	100	1.5	12.75	_100	0.625
1N71	SYL.	40	200	60	1000	1.0	15	30	0.30
1N73	CBSH	60	60	22.5	100	1.3	15	_10	0.05
1N73 G9	GE			22.5	100	1.3— 1.7	15	_10	0.05
1N74	СВЅН	60	60	22.5	100	1.2	15	—10	0.05
1N74 G9A	GE			22.5	100	1.2—  1.8	15	10	0.05
СК709	RAY	60	_	50					
CK711	RAY	80	_	35		-			
CBSH CBS-Hytron GE General Electric					RAY. Raytheon SYL. Sylvania				
SPECIAL 4-DIODE BALANCED GERMANIUM MODULATORS.									

ployed in wave analyzers. The input transformer has been replaced by a phase inverter,  $V_1$ , which converts the single-ended modulating-voltage  $(f_1)$  input to out-of-phase input for the grids of  $V_2$  and  $V_3$ . The carrier  $(f_2)$  is applied, through transformer  $T_1$ , to the cathodes of  $V_2$  and  $V_3$  in parallel.

In this circuit, the frequency responds to  $T_2$  usually is such that the modulating frequency,  $f_1$ , could not be passed, even if stage feedthrough should chance to occur. In wave analyzer applications, the modulator output is presented to the input of a highly-selective bandpass filter for transmitting only one sideband.

Efficient operation of tube-type balanced modulators requires tubes matched for transconductance (and often for plate and screen currents, as well), and closely regulating electrode voltages. Center-tapped transformers are designed for close balance between sections and for equal small capacitances between the extremities of primary and secondary windings. Diode-Type Balanced Modulators

The non-linear volt-ampere characteristic of semiconductor diodes suits these simple 2-terminal components to the function of modulation. Connecting four such *matched* diodes into a suitable symmetrical circuit yields a simple, compact balanced modulator requiring no bias and filament potentials.

Originally, copper oxide rectifiers were used for this purpose but they are limited with respect both to maximum signal voltage and frequency. While copper oxide modulators still are preferred for certain lowfrequency applications, they have been supplanted largely by germanium diodes because of the superior voltage and frequency characteristics of the latter.

Figure 3 shows three common diode-type balanced modulator circuits. While the particular mode of operation is somewhat different for each circuit, the modulating action is about equal in the end result. The balanced arrangement of matched diodes results in suppression of the carrier

and transmission of upper and lower sidebands. The circuits in Figures 3(A) and 3(B) permit a common connection between modulation input and sideband output, and this common point can be grounded. This is a desirable feature in some instances. The circuit of Figure 3(C) provides no such path. However, Figure 3(C) allows both the carrier and modulation voltage sources to have a common connection, which can be grounded, while Figures 3(A) and 3(B) necessitate floating the carrier source. Choice of circuit therefore will be governed by prevailing installation and operating requirements.

Observe that the diodes,  $D_1$ ,  $D_2$ ,  $D_3$ , and  $D_4$ , in Figures 3(A) and 3(B) are connected together in a conventional bridge circuit. The different arrangement in Figure 3(C), however, is a configuration known as a ring modulator or double-balanced modulator.

The diode-type modulator has been described as a carrier-operated switch. In this respect, it may be considered to open, short, or reverse



consists of diodes  $D_1$  and  $D_2$  and the two halves of balancing potentiometer R. The carrier voltage switches the lower end of  $L_1$  alternately to ground at the carrier rate.

In the double-tuned output transformer,  $L_1C_1$  and  $L_2C_2$  each is tuned to resonate at the output frequency (sideband). For efficient operation of the circuit, the impedance  $Z_1$  of the modulating source must be low for the carrier frequency, and the output impedance  $Z_2$  must be low for the modulating frequency. Capacitance to ground must be kept as low as possible at the junction of  $D_1$  and  $D_2$ . Capacitance between the transformer windings,  $L_1$  and  $L_2$ , likewise must be minimized.

Intermodulation products generated in this circuit are said to be lower than 60 db when the carrier amplitude is 3 volts and the modulating signal amplitude 0.1 v.

#### Balanced Modulator as DC-AC Inverter

If a d. c. voltage is substituted for the carrier in a balanced modulator, the output will have the frequency of the carrier and will be proportional to the applied d. c. volt-



age. This action provides a simple means for changing small d. c. potentials to proportionate a. c. voltages for measurement or control. Thus, small d. c. voltages may be measured with a suitably calibrated a. c. millivoltmeter.

The diode-type balanced modulator is an attractive DC-AC inverter for low power levels, since it uses no tubes, requires no d. c. supply voltages, and is ready for instant operation. The semiconductor-diode type offers the additional advantage of freedom from contact potentials.

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