march 1960 the institute of radio engineers

Proceedings of the IRE

INTERNATIONAL CONVENTION IRE





MARCH 21-24

Waldorf-Astoria Hotel and New York Coliseum











UTC NEW DO-TAND DI-TSERIES Revolutionary transistor transformers hermetically sealed to MIL-T-27A Specifications.

UTC DO-T and DI-T transistor transformers provide unprecedented power handling capacity and reliability coupled with extremely small size. Comparative performance with other available products of similar size are shown in the curves (based on setting output power at 1 KC, then maintaining same input level over frequency range). The new expanded series of units cover virtually every transistor application.



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And Special Units to Your Specifications

DO-T No.	MIL Type	Application	Pri. Imp.	O.C. Ma.‡ in Pri.	Sec. Imp.	Pri. Res. 00-T	Pri. Res. 01-T	Level Mw.	OI-T No,
DO-T1	TF4RX13YY	Interstage	20,000	.5	800 1200	850	815	50	DI-T1
D0-T2	TF4RX17YY	Output	500 600	3	50 60	60	65	100	01-12
DO-T3	TF4RX13YY	Output	1000	3	50 60	115	110	100	01-T3
DO-T4	TF4RX17YY	Output	600	3	3.2	60		100	
00-T5	TF4RX13YY	Output	1200	2	3.2	115	110	100	01-T5
DO-T6	TF4RX13YY	Dutput	10,000	1	3.2	790		100	
DO-17	TF4RX16YY	Input	200,000	0	1000	8500		25	
DO-T8	TF4RX20YY	Reactor 3.5 Hys. @ 21	Ma. DC, 1 Hy. @	5 Ma. DC		630			
	TF4RX20YY	Reactor 2.5 Hys. @ 2	Ma. DC, .9 Hy. (0	4 Ma. DC	;		630		DI-T8
DO-T9	TF4RX13YY	Dutput or driver	10,000 12,000	1	500 CT 600 CT	800	870	100	DI-T9
D0-T10	TF4RX13YY	Driver	10,000 12,000	1	1200 CT 1500 CT	800	870	100	DI-T10
DO-T11	TF4RX13YY	Driver	10,000 12,000	1	2000 CT 2500 CT	800	870	100	D1-T11
DO-T12	TF4RX17YY	Single or PP output	150 C 200 C	T 10 T 10	12 16	11		500	
DO-T13	TF4RX17YY	Single or PP output	300 C 400 C	T 7 T 7	12 16	20		500	
DO-T14	TF4RX17YY	Single or PP output	600 C 800 C	T 5 T 5	12 16	43		500	
00-T15	TF4RX17YY	Single or PP output	800 C 1070 C	T 4 T 4	12 16	51	_	500	
DO-T16	TF4RX13YY	Single or PP output	1000 C 1330 C	T 3.5 T 3.5	12 16	71		500	-
00-T17	TF4RX13YY	Single or PP output	1500 C 2000 C	T 3 T 3	12 16	108		500	
DO-T18	TF4RX13YY	Single or PP output	7500 C 10,000 C	T 1 T 1	12	505		500	
DO-T19	TF4RX17YY	Output to line	300 C	T 7	600	19	20	500	DI-T19
DG-T20	TF4RX17YY	Output or line to line	500 C	T 5.5	600	31	32	500	01-120
DO-T21	TF4RX17YY	Output to line	900 C	1 4	600	53	53	500	01.121
00-122	TF4RX13YY	Output to line	1500 0	1 3 T 5	900 CT	00	915	100	DI-T22
00-123	TEADALCAN	Interstage	30,000 C	T .5	1200 CT	9500	010	25	
00-124	TEADYLONY	chopper service)	200.000 0	T 1	1500 CT	800	870	100	DI-T25
00-125	TEARNOONN	Interstage	12,000 C	T 1	1800 CT	2100	070		
00-126	TF4RX2011	Reactor 6 Hy. @ 2 Md.	DC 12 Hy @	A Ma DC		2100	2300		DI-T26
00 127	TEARX2011	Reactor 1 25 Hy (c) 2 M	a. DC, 1.2 mj. (a)	1 Ma DC		100	2300		
00.127	TEARX20YY	Reactor 9 Hy to 2 Ma	DC 5 Hy (a 6	Ma. DC			105		DI-T27
DO-T28	TEARX20YY	Reactor 3 Hy a 4 Ma	DC. 15 Hy. (a)	20 Ma. DC		25			
_	TF4RX20YY	Reactor 1 Hy. a 4 Ma.	DC, .08 Hy. @	10 Ma. DC			25		01-T28
DO-T29	TF4RX17YY	Single or PP output	120 C 150 C	T 10 T 10	3.2	10		500	
DO-T30	TF4RX17YY	Single or PP output	320 C 400 C	T 7 T 7	3.2 4	20		500	
DO-T31	TF4RX17YY	Single or PP output	640 C 800 C	T 5 T 5	3.2	43		500	
DO-T32	TF4RX17YY	Single or PP output	800 C 1,000 C	T 4 T 4	3.2 4	51		500	
DO-T33	TF4RX13YY	Single or PP output	1,060 C 1,330 C	T 3.5 T 3.5	3.2 4	71		500	
DO-T34	TF4RX13YY	Single or PP output	1,600 C 2,000 C	T 3 T 3	3.2 4	109		500	
00-T35	TF4RX13YY	Single or PP output	8,000 C 10,000 C	T 1 T 1	3.2 4	505		500	
DO-T36	TF4RX13YY	Isol. or Interstage	10,000 C	T 1	10000 CT	950	970	500	DI-T36
DO-TSH	Drawn Hiper	malloy shield and cover	for DO-T's, pr	ovides 25	to 30 db	shieldin	g, for	DI-T's	DI-TSH

‡DCMA shown is for single ended useage (under 5% distortion—100MM—1KC) . . for push pull, DCMA can be any balanced value taken by .5W transistors (under 5% distortion—500MW—1KC) *00-T units have been designed for transistor application only . . . not for vacuum tube service. Pats. Pend. *00-T units have been designed for transistor application only . . . not for vacuum tube service. Pats. Pend. *00-T units have been designed for transistor application only . . . not for vacuum tube service. Pats. Pend. *00-T units have been designed for transistor application only . . . not for vacuum tube service.

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March, 1960

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MARCH

Proceedings of the IRE

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Proceedings of the IRE[®]

continued

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IRE NEW



at the IRE NATIONAL CONVENTION and RADIO-ENGINEERING SHOW!

It doesn't matter how you manage it - by starting at the fourth floor with Production Items, on to the third floor for Systems and Instruments, then down to Two and One for Components – or the reverse – what does matter is that you see ALL there is to see at the IRE National Convention and Radio-Engineering Show at the New York Coliseum, March 21-24. You could even take in one floor a day! Remember, there are 4 BIG FLOORS... and 4 BIG DAYS...so, plan your trips to the Coliseum so that you don't miss anything.

The opportunity to see SO MUCH that's NEW in the radio-engineering field comes but once a year with this giant IRE National Convention and Radio-Engineering Show. Be UP on your field with a thorough knowledge of the displays and exhibits that will be shown as NEW IDEAS in RADIO-ELEC-TRONICS, from the top fourth floor to the bottom first floor, at the New York Coliseum!



7

MARCH 21, 22, 23, 24

Waldorf-Astoria Hotel

The IRE NATIONAL CONVENTION | The RADIO ENGINEERING SHOW Coliseum, New York City

The Institute of Radio Engineers • 1 East 79th St., New York 21, N.Y.

PROCEEDINGS OF THE IRE March, 1960

3A



The series of advertisements which we recently ran in this space on the subject of Reliability was very well received, judging from the letters which were sent to us. This month W. A. Jasson, a member of the Reliability Group that is assigned to space projects, describes another facet of reliability engineering.

Step-Stress Method for Comparative and Survival Evaluation

We who stringle with weapons systems development and allied anxieties have a number of problems in common. One of the most difficult of these is the selection of component parts and circuits such that the probability of system survival (reliability) is maximized. This problem also calls attention to another: how to evaluate the probability of system survival in the first place. Since few answers are in prior data or the literature, we can resort to an economical and meaningful experimental procedure: step-stress testing. This can yield significant results with small samples.

In the step-stress technique of comparative evaluation and survival testing, malfunctions are induced in the test subject. The procedure is to select and apply stresses related to the system mission of such type and first-level magnitude as are likely to be withstood by all parts in our tests. After applying such a first-level stress, we take pertinent measurements of its effects. We then raise this stress to a pre-determined higher level and take measurements again. The process continues as stresses are increased in steps and the effects observed and analyzed. When the number of malfunctions caused increases to a point considered adequate (or the maximum practicable) for the purposes of the test, step-stressing ends (assuming all relevant stresses have been stepped) and we are ready for analysis of results.

Stresses applied in these step-stress tests can be divided into two categories: environmental and electrical. The sequence of (and within) these categories must be decided in the light of system and parts requirements. For a missile-borne system, it makes sense to start with environmental stresses, since these occur in storage and handling, and they attend the launching, usually preceding electrical stresses. First of the environmental stresses to apply are shock and vibration, which, although not always encountered first, are apt to yield more information than the others.

For the mathematical evaluation of environmental step-stress test data, most value will be found in significance tests which are often supplemented by regression analyses (particularly comparing prestress and post-stress values) and correlation. For life test data, the most useful calculations involve significance tests, hazard rates (failures/million hours in a convenient unit), and mean-time-to-failure (the reciprocal of the hazard rate). If stepstress test analyses are to be strictly valid, we must assume that the probabilities of failure of the units under test do not permute as stress level changes (see Figure 1). If this assumption is false, a part having a high failure rate at a high stress level may be stronger than its competitors are at more relevant stress levels. We must also be sure that the stress types are pertinent to the system mission and conditions, and that each stress (regardless of level) is applied in accordance with all environmental and electrical test specifications applicable for single-level tests. One should also consider the relative importance of



fatigue and one-shot failure factors in the actual system application. The selection of stress levels should reflect the balance of significance of these two factors. Thus, if fatigue is the over-riding consideration, it would tend to show up better if the earlier stress levels are much closer together than would be desirable were one-shot failure the paramount problem.

Let us consider the comparative evaluation of three manufacturers' versions of a particular type of silicon switching tran-This transistor is to be used in a sistor missile-borne system. Suppose we have determined the limit values of those transistor characteristics necessary for proper oper-ation of the circuit. These characteristics would probably include pulse DC beta (hFE), collector cut-off current (ICBO), switching time, and saturation resistance. We can class as failures those transistors which change such that their characteristics move outside these limit values. Naturally, failures of this nature must be considered in addition to any catastrophic failures (shorts, opens, or characteristics values approaching these conditions). Since the system is to be missile-borne, we shall tackle shock and vibration stressing first. Figure 2 shows both the shock stress steps used and the corresponding number of cumulative failures resulting. The shock stress steps served as the basis for proportionate increments in vibration level. Whereas the results at the highest and lowest shock levels may not be significant, the large differences at the middle levels indicate that vendor A is superior.

In general, note that in running tests such as these, the usual procedure would be to alternate shock and vibration stresses. The shock stresses themselves should be applied in both directions along two or three axes in the usual manner. The application of vibration and all other stresses



should also be as usual. If a particular step-stress level should cause all parts to fail, the test may have to be re-run using a lower level.

Other environmental tests also lend themselves to step-stressing, particularly thermal cycling, wherein we widen in steps the temperature limits of the cycles. An alternative, or a supplement, to this would be thermal shock step-stressing, which is more severe.

Turning now from environmental to electrical step-stress tests, their basic principle is very much the same. In electrical step-stressing, we step up, on a given time schedule, the electrical load upon the devices under test. A load so stepped may be in terms of power dissipations, applied voltages or currents, or both, depending upon (1) applicable component-part specifications of the manufacturer or of the project, (2) circuit or system specifications, or (3) knowledge of failure points of the part or circuit. The mathematical analysis of electrical step-stress test results and life test as discussed above.

From all the foregoing, perhaps it can be seen that step-stressing is really a logical extension of single-level stress tests aimed at providing survival information for analysis or comparison of entities with respect to a particular electronic system. The step-stress technique yields accelerated failures. For any effective reliability effort, which normally includes comparative evaluation and/or survival testing, proficiency in the art and science of applying stepstress techniques is rapidly becoming a prerequisite to success.

A complete bound set of our third series of AlL Monographs is available on request. Write to Harold Hechtman at AlL for your set.

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HOW TO SELECT HIGH RELIABILITY CAPACITORS

At one time Sprague Electric was the only manufacturer offering true high reliability capacitors. The buyer had no problem. But today there are many manufacturers who claim that their capacitors meet high reliability standards. Some are even so bold as to claim that theirs are *the most reliable*.

☆

Check the record before you choose

The only sound approach to evaluate these claims is to investigate the *reliability record* achieved by each of the companies under consideration. Remember, it takes test data to establish the reliability of a product. Claims are not enough.

Now let's look at the record

Sprague Electric can substantiate its claim that its HYREL® Q Capacitors are "the most reliable capacitors made" with the most extensive test data available in the entire electronic industry. The performance of HYREL Q Capacitors is virtually impossible to surpass...now and for some years to come.

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On accelerated life tests the failure rate of HYREL Q Capacitors has been less than 0.05%, after more than 16 million unit hours accumulated on tests of 250 hours at 140% rated voltage, 125 C. On high frequency vibration tests, there hasn't been a single failure in the more than 50,000 units tested. On seal, moisture resistance, and temperature cycling and immersion tests, the failure rate has been less than 0.1%.

Such performance from production line capacitors can only be achieved through the most intensive (and expensive) kind of reliability program—in design and development, in production engineering, in manufacturing facilities, in testing intensity and extensity—all of which should be investigated thoroughly.

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For complete facts and figures on HYREL Q Capacitors, call your Sprague District Office or Representative, or write for HYREL Bulletin 2900A and Specification PV-100A to Technical Literature Section, Sprague Electric Company, 23.3 Marshall St., North Adams, Massachusetts.

PROCEEDINGS OF THE IRE March,

JEE U. March, 1960

SEE US AT THE IRE SHOW-BOOTHS 2416-2424



HE X-RAYS WOOD...

to help make telephone poles last longer





Chemist Jack Wright developed the use of this X-ray fluorescence machine for testing the concentration of preservatives in wood. Here he bombards a boring from a test telephone pole with X-rays.

This Bell Labs chemist is using a fast, new technique for measuring the concentration of fungus-killing preservative in telephone poles.

A boring from a test pole is bombarded with X-rays. The preservative—pentachlorophenol—converts some of the incoming X-rays to new ones of different and characteristic wave length. These new rays are isolated and sent into a radiation counter which registers their intensity. The intensity in turn reveals the concentration of preservative.

Bell Laboratories chemists must test thousands of wood specimens annually in their research to make telephone poles last longer. Seeking a faster test, they explored the possibility of X-ray fluorescence—a technique developed originally for metallurgy. For the first time, this technique was applied to wood. Result: A wood specimen check in just two minutes—at least 15 times faster than before possible with the conventional microchemical analysis.

Bell Labs scientists must remain alert to *all* ways of improving telephone service. They must create radically new technology or improve what already exists. Here, they devised a way to speed research in one of telephony's oldest and most important arts—that of wood preservation.

Nature still grows the best telephone poles. There are over 21 million wooden poles in the Bell System. They require no painting, scraping or cleaning: can be nailed, drilled, cut, sawed and climbed like no other material. Scientific wood preservation cuts telephone costs, conserves valuable timber acres.



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TR24R	24	0-2	5 x 41/4 x 63%	160.
TR32R	32	0-2	5 x 41/a x 63/a	160
TR6-32R	6-32**	0.2	5 x 41/4 x 63/2	185

* Prices FOB Cedar Grove, subject to change without notice

** Selectable voltages at 6, 12, 18, 24 or 32 VDC

In addition to models listed, units can be supplied to meet special military or commercial requirements. Write for quotations on special types.

For further details send for catalogue #118.



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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



As a service both to Members and the industry, we will endeavor to record in this column each month those meetings of IRE, its sections and professional groups which include exhibits.

- 1

March 21.24 1960

- IRE 1960 International Convention and Engineering Show, Waldorf-Astoria Hotel and New York Coliseum. New York, N.Y.
- Exhibits: Mr. William C. Copp. Institute of Radio Engineers, 72 West 45th St., New York 36, N.Y.

April 3-8, 1960

- Sixth Nuclear Congress, New York Coliseum, New York, N.Y. Exhibits: Mr. F. M. Howell, c/o EJC, 29 W. 39th St., New York 18, N.Y.

April 20.22, 1960

- SWIRECO, Southwestern IRE Regional Conference & Electronics Show, Shamrock-Hilton Hotel. Houston, Texas.
- Exhibits: Mr. A. D. Seixas, SWIRECO, P.O. Box 22331, Houston, Texas.

May 2-4 1960

- National Aeronautical Electronics Conference, Dayton Biltmore Hotel, Dayton, Ohio.
- Exhibits: Mr. Edward M. Lisowski, General Precision Lab., Inc., Suite 452, 333 West First St., Dayton 2, Ohio.

May 2-6, 1960

Western Joint Computer Conference, Fairmont Hotel, San Francisco, Calif. Exhibits: Mr. H. K. Farrar, Pacific Tel. & Tel. Co., 140 New Montgomery St., San Francisco 5, Calif.

- May 24.26, 1960
 - Seventh Regional Technical Conference & Trade Show, Olympic Hotel. Seattle, Wash.
 - Exhibits: Mr. Rush Drake, 1806 Bush Place, Seattle 44, Wash.
- May 24-26, 1960
 - Armed Forces Communications & **Electronies Association Convention** and Exhibit, Sheraton-Park Hotel. Washington, D.C.
 - Exhibits: Mr. William C. Copp, 72 West 45th St., New York 36, N.Y.
- June 20-21, 1960
- Chicago Spring Conference on Broadcast and Television Receivers, Graemere Hotel, Chicago, Ill.
- Exhibits: Mr. Stanley Hopper, Zenith Radio Corp., 6001 W. Dickens Ave., Chicago 39, Ill.

June 27-29, 1960

- National Convention on Military Electronics, Sheraton-Park Hotel, Washington, D.C.
- Exhibits: Mr. L. David Whitelock, Bu-Ships, Electronics Div., Dept. of Navy, Washington, D.C.

(Continued on page 10.4)

World Radio History



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SPECIFICATIONS

VARIABLE FREQUENCY RANGES: .5-12 mc, .1-12 mc, 10 kc-12 mc, 10-220 mc (9 bands)

FIXED FREQUENCIES: Up to max, of 8 center frequencies (20 kc to 12 mc—Customer selected)

AUDIO RANGE: 200 cps to 20 kc.

- SWEEP WIDTHS: Selected for maximum stability 1-10 mc on .5-12 mc band; .2-2 mc on .1-12 mc band; 20-200 kc on 10 kc-12 mc band; 6% to 60% of center freq. to 50 mc and 3 mc to 30 mc above 50 mc on 10-220 mc bands. 2-20 kc on fixed frequencies and audio range.
- OUTPUT LEVEL: Continuously variable from 1 volt rms down to 65 db below 1 volt, $\pm 5\%$ over widest sweep. AGC. Audio range: variable .5-1 volt rms.
- 1MPEDANCE: 70 ohms nominal (50 ohms on request). Audio range: 600 ohms.
- SWEEP OUTPUT and REPETITION RATES: Sawtooth for horizontal deflection of oscilloscope. Approx. 7 volts peak to peak—Output Impedance 1000 ohms nom.; fixed 60 cps, line locked; fixed 30 cps. logarithmic (tor audio and video application) 3 cont. var. ranges—.2-1 cps, 1-5 cps, 5-30 cps.
- MARKERS: Swept signal available for operation of *Vari-Marker* SKV Generator.
 - Optional Internal Markers, Lunned number of sharp, crystalcontrolled pulse-type markers a customer specified frequencies can be provided. Please inquire before ordering.
- POWER SUPPLY: Input approx. 220 Watts, 117v (±10%), 50-60 cps. B+ electronic regulation.
- PRICE: \$995.00 f.o.b. factory. Fixed freq bands add \$17.00 per band.

The wide range of frequency and repetition rate in the Ligna-Sweep Model SKV make it ideally suited for alignment and testing of a wide variety of electronic instruments—audio amplifiers. filters. communication receivers. radar IF channels. TV receivers and transmitters.

The unit is stable and carefully shielded and filtered to prevent spurious signals on beat frequency video bands. A wide range of sweep repetition rates makes viewing easy on conventional oscilloscopes. Low repetition rates used with long persistence screens permit study of high Q circuitry. LF limits of band circuits and observation of the "ring" characteristics of tuned circuits.



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CApital 6-4000



Recently installed on the atomic submarine SK1PJACK (SSN585), the Westinghouse Electric AN/WRT-2 SSB Transmitter is now standard Navy equipment.

Single sideband signals are generated in the AN/WRT-2 by the selective filter method employing Hermes 2MUB and 2MLB Crystal Filters. These 2.0 Mc Crystal Filters not only offer all the basic advantages of the filter SSB generation method, but reduce the number of heterodyning stages required to translate the modulated signal to the required output frequency. The attendant decrease in unwanted signal generation results in a cleaner signal. The AN/WRT-2 is also a more reliable transmitter because fewer components are used.

In addition to the 2.0 Mc Crystal Filters, Hermes has also supplied SSB units at 87 Kc, 100 Kc, 137 Kc, 1.4 Mc, 1.75 Mc, 3.2 Mc, 6 Mc, 8 Mc, 10 Mc and 16 Mc. These Crystal Filters are presently installed in airborne HF, mobile VHF and point to point UHF SSB systems.

Whether your selectivity problems are in transmission or reception, AM or FM, mobile or fixed equipment, you can call on Hermes engineering specialists to assist in the design of circuitry and the selection of filter characteristics best suited to your needs. Write for Crystal Filter Short Form Catalog.

A limited number of opportunities are available to experienced circuit designers. Send résumé to Dr. D. I. Kosowsky.





(Continued from page 8A)

August 23-26, 1960

- WESCON, Western Electronic Show and Convention. Ambassador Hotel & Memorial Sports Arena, Los Angeles, Calif.
- Exhibits: Mr. Don Larson, WESCON, 1435 LaCienega Blvd., Los Angeles, Calif.
- September 19-21, 1960
 - National Symposium on Space Electronics & Telemetry, Shoreham Hotel, Washington, D.C.
 - Exhibits: John Leslie Whitlock Associates, 6044 Ninth St., North, Arlington 5, Va.
- October 3.5, 1960
- Sixth National Communications Symposium, Hotel Utica & Utica Municipal Auditorium, Utica, N.Y. Exhibits: Mr. B. F. Bischoff, 19 West-
- Exhibits: Mr. R. E. Bischoff, 19 Westminster Road, Utica, N.Y.
- October 10-12, 1960
- National Electronics Conference, Hotel Sherman, Chicago, Ill.
- Exhibits: Mr. Arthur H. Streich, National Electronics Conference, 184 E. Randolph St., Chicago, Ill.
- October 24-26, 1960
 - East Coast Aeronautical & Navigational Electronics Conference, Lord Baltimore Hotel & 7th Regiment Armory, Baltimore, Md.
 - Exhibits: Mr. R. L. Pigeon, Westinghouse Electric Corp., Air Arm Div., P.O. Box 746, Baltimore, Md.
- Oct. 31-Nov. 2, 1960
 - 13th Annual Conference on Electrical Techniques in Medicine & Biology, Sheraton-Park Hotel, Washington, D.C.
 - Exhibits: Mr. Lewis Winner, 152 West 42nd St., New York 36, N.Y.
- November 14-16, 1960
- Mid-America Electronics Convention (MAECON), Municipal Auditorium, Kansas City, Mo.
- Exhibits: Mr. John V. Parks, Bendix Aviation Corp., P.O. Box 1159, Kansas City 41, Mo.
- November 15-17, 1960
- Northeast Electronics Research & Engincering Meeting (NEREM), Boston Commonwealth Armory, Boston, Mass.
- Exhibits: Miss Shirley Whitcher, IRE Boston Office, 73 Tremont St., Boston, Mass.

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Note on Professional Group Meetings: Some of the Professional Groups conduct meetings at which there are exhibits. Working committeemen on these groups are asked to send advance data to this column for publicity information. You may address these notices to the Advertising Department and of course listings are free to IRE Professional Groups.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

The New Ramo-Wooldridge Laboratories in Canoga Park

...an environment dedicated to technological research and development

The new Ramo-Wooldridge Laboratories in Canoga Park, California, will provide an excellent environment for scientists and engineers engaged in technological research and development. Because of the high degree of scientific and engineering effort involved in Ramo-Wooldridge programs, technically trained people are assigned a more dominant role in the management of the organization than is customary.

The ninety-acre landscaped site, with modern buildings grouped around a central mall, contributes to the academic environment necessary for creative work. The new Laboratories will be the West Coast headquarters of Thompson Ramo Wooldridge Inc. as well as house the Ramo-Wooldridge division of TRW.

The Ramo-Wooldridge Laboratories are engaged in the broad fields of electronic systems technology, computers, and data processing. Outstanding opportunities exist for scientists and engineers.

For specific information on current openings write to Mr. D. L. Pyke.





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The name TRIPLETT has been on instruments of our manufacture for more than 55 years, and is regarded as a symbol of customer satisfaction to industrials and distributors in all parts of the world. Our instruments can be built to customer

specifications or provided from our large stocks of standard ranges in hundreds of sizes and types. We also carry in stock many semi-finished movements which can be converted readily to special customer needs.



World Radio History

CURRENT IRE STATISTICS

(As of January 31, 1960) Membership—79,514 Sections*—105 Subsections*—27 Professional Groups*—28 Professional Group Chapters—261 Student Branches†—184

* See this issue for a list. † See October, 1959 issue for a list.

Calendar of Coming Events and Authors' Deadlines*

1960

- IRE National Conv., N. Y. Coliseum and Waldorf-Astoria Hotel, New York, N. Y., Mar. 21-24.
- First Natl. Symp. on Human Factors in Electronics, BTL Aud., New York, N. Y., Mar. 24-25.
- Scintillation Counter Symp., Washington, D.C., Mar.
- 6th Nuclear Congress, N. Y. Coliseum. New York, N. Y., Apr. 4-8.
- 14th Spring Tech. Conf., Cincinnati, Ohio, Apr. 12-13.
- Conf. on Automatic Tech., Sheraton-Cleveland Hotel, Cleveland, Ohio, Apr. 18-19.
- Int'l Symp. on Active Networks and Feedback Systems, Engrg. Soc. Bldg. Auditorium, New York, N. Y., Apr. 19-21.
- Int'l Symp. on Active Networks and Feedback Systems, Polytechnic Inst. of Brooklyn, Brooklyn, N. Y., Apr. 19-21.
- 1960 SWIRECO (Southwestern IRE Regional Conf. and Electronics Show), simultaneously with the Nat'l. PGME Conf., Houston, Texas, Apr. 20-22.
- Natl. Aeronautical Electronics Conf., Biltmore and Miami-Pick Hotels, Dayton, Ohio, May 2-4.
- URSI-IRE Spring Mtg., Sheraton Park Hotel and NBS, Washington, D. C., May 2-5.
- Western Joint Computer Conf., San Francisco, Calif., May 2-6.
- PGMTT Natl. Symp., San Diego, Calif., May 9-11.
- Electronic Components Conf., Hotel Washington, Washington, D. C., May 10-12.
- 7th Reg. Tech. Conf. & Trade Show, Olympic Hotel, Seattle, Wash., May 24-26.
- 6th Radar Symp., Ann Arbor, Mich., June 1-3.
- Conf. on Standards and Electronic Measurements, NBS Boulder Labs., Boulder, Colo., June 22-24.

* DL = Deadline for submitting abstracts.

(Continued on page 15A)

Chicago Spring Conference To Be Held in June

The Chicago Spring Conference (C.S.C.) on Broadcast and Television Receivers will be held for the first time at the Grasmere Hotel on June 20 and 21, 1960. Sponsors are the Professional Group on Broadcast and Television Receivers and the Chicago Section of the IRE.

For the most part the C.S.C. will concern itself with entertainment consumer fields. It will also continue the television aspect of the Spring Technical Conference in Cincinnati, and will run concurrently with the June Furniture Show.

Technical papers on Broadcast and Television Receivers and related fields are requested. Further information may be obtained by contacting Pieter Fockens, C.S.C. Chairman, Zenith Radio Corp., 6001 W. Dickens, Chicago 39, Ill.; Jack E. Bridges, C.S.C. Papers Chairman, Warwick Manufacturing Co., 7300 N. Lehigh, Chicago, Ill.; William G. Henke, C.S.C. Publicity Chairman, Admiral Corp., 3800 Cortland Ave., Chicago 47, Ill.; or Stanley Hopper, C.S.C. Exhibits Chairman, Zenith Radio Corp., 6001 W. Dickens, Chicago 39, Ill.

PGHFE WILL HOLD Annual Symposium

The Professional Group on Human Factors in Electronics (PGHFE) will hold its First Annual Symposium on Human Factors in Electronics in New York, N. Y. in the auditorium of Bell Telephone Laboratories, 463 West St. (use Bethune Street entrance), on the evening of March 24 from 7:30 p.M. to 10:00 p.M. and all day on March 25 from 9:30 A.M. to 5:30 p.M.

On the evening of March 24 there will be a symposium on the topic: "Human Factors in Electronics—a Progress Report from Industry." A selected panel of speakers from various industrial organizations will consider the questions:

- 1) How and to what extent is industry using human factors engineering?
- 2) What can the PGHFE do to support the work of human factors engineers in the electronic and related industries?

On March 25 a series of papers will be presented on topics such as: Theory of Man-Machine Systems, Effect of Environment on Human Operation of Electronic Equipment, Evaluation of Human Factors Design of Systems, and so forth.

Registration fees are \$2.00 per person for IRE members and \$3.00 per person for nonmembers. Attendance is limited; advance registration is recommended. Application forms may be obtained from:

K. G. Van Wynen, Chairman Local Arrangements Committee Bell Telephone Laboratories, Room 628A 463 West Street New York, N. Y.

Computer Federation Formed by Eleven Nations

Computers and information processing assumed global importance in January with the birth of a new international body dedicated to this fast-growing technology. Eleven nations have ratified the statutes of the International Federation of Information Processing Societies, which for the first time will provide a common meeting ground for computer experts from all over the world. Until now, many countries, including the United States, have had their own professional computer societies, but these groups have had no permanent, formal means of meeting and exchanging ideas. The need for better communication in the field of information processing is becoming increasingly important as all nations come to appreciate the vital role computers play in man's daily life.

The movement to form the new Federation was a direct result of the first International Conference on Information Processing, sponsored by the United Nations Educational, Scientific and Cultural Organization (UNESCO) and held in Paris last June. As a result a provisional bureau for the International Federation was established, with Isaac L. Auerbach, president of Auerbach Electronics Corp., Narberth, Pa., named provisional chairman. Mr. Auerbach represents the National Joint Computer Committee of the U.S. and was U. S. consultant to UNESCO for the Paris conference. Also named to the provisional committee were Professor A. A. Dorodnicyn, of the U.S.S.R., and A. van Wijngaarden, of The Netherlands, vice-chairmen, and J. A. Mussard, of UNESCO, secretary.

The countries whose national computer technical societies have ratified the statutes include Canada, Denmark, Finland, France, Germany, The Netherlands, Spain, Sweden, Switzerland, the United Kingdom, and the United States. In addition, Belgium, Israel, and Japan are forming national computer societies to qualify for membership.

It is expected that the first meeting of the IFIPS council later this year will result in plans for a second International Conference on Information Processing with an associated technical exhibit in 1963.

Membership in the IFIPS now includes: Computing and Data Processing Society of Canada (Canada), Danish Academy of Technical Sciences (Denmark), Finnish National Committee for Information Processing (Finland), Association Française de Calcul (France), Deutsche Arbeitsgemeinschaft für Rechen-Anlagen (West Germany), Nederlands Rekenmachine Genootschap (Netherlands), Instituto de Electricidad y Automatica (Spain), Swedish Society for Information Processing (Sweden), Swiss Federation of Automatic Control (Switzerland), British Computer Society (United Kingdom), and the National Joint Computer Committee (U.S.A.).

The National Joint Computer Committee comprises representatives of the Institute of Radio Engineers, the Association for Computing Machinery, and the American Institute of Electrical Engineers.

<u>Worl</u>d Radio History

INTERNATIONAL CONGRESS TO BE HELD IN ROME

The "VIIa Rassegna Internazionale Elettronica Nucleare e della Cinematografia" will be held at the Palazzo dei Congressi, Rome, Italy, from June 15 to 29. 1960, and will include an International Technological Exhibition where all the newest and most interesting projects in the fields of Electronics, Nuclear Energy and Cinematography are being shown; and International Scientific Congresses in which Scientists from the world over take part by submitting papers and/or giving lectures. National official delegations will also be present.

The Scientific Congresses and the Exhibition are both organized to represent a yearly synthesis of the progress internationally achieved in the above-mentioned fields.

The Provisional Programs of the Scientific Congresses are the following:

Electronics

- 1) Modern amplifiers (parametric and molecular amplifiers).
- 2) Spatial radiocommunications.
- 3) Problems connected with broadcasting on bands IV and V.
- 4) Electronic computers. Collation and processing of data for research operation.
- 5) General survey of progress in electropics

Cinematography

- a) Long range cinematography and photography.
- b) Influence of shooting and screening sizes on the development of cinematography.

Nuclear Energy

Still to be completed.

The Proceedings of the Scientific Congresses of 1959 will be released shortly.

A monograph in five languages by the title of "L'Industria Nucleare Italiana" has been edited by the "Rassegna" in collaboration with the Ministry and the Institute for Foreign Trade, where all Italian Nuclear Industries are listed together with an illustrated presentation of their products. This Monograph may be obtained at the price of Italian Lira 2000 per copy from the "Segreteria Rassegna Internazionale Elettronica Nucleare e della Cinematografia, Via della Scrofa 14, Roma."

LONG ISLAND SECTION To Honor New Fellows

The Long Island Section of the IRE will honor the newly elected Fellows of this section and also newly elected overseas Fellows at a Fellow award presentation on Sunday, March 20, at 4:30 P.M. in the Garden City Hotel, Garden City, N. Y.

The newly elected Fellows from the Long Island Section are Dr. C. E. Dean, Dr. R. K. Hellmann, Prof. W. A. Lynch, D. S. Rau, and W. E. Tolles. A reception for members and guests will follow.

ARMY MARS NET APPOINTS Associate Directors

J. P. Hoffman, Information Officer and Frederic H. Dickson, Chief of the Radio Propagation Agency of the U.S. Army Signal Corps at Ft. Monmouth, have been appointed associate net directors of the First U. S. Army MARS SSB Technical Net.

Mr. Hoffman and Mr. Dickson will make arrangements for originating one speaker each month from among the Electronic Scientists and Engineers at Ft. Monmouth.

The net can be heard each Wednesday evening at 9 P.M. EST on 4030 kc upper sideband. The schedule for March includes:

- March 2 "Transistorized Test Equipment for the Amateur Radio Station,' R. W. Gunderson, Editor, Braille Technical Press, New York, N. Y.
- March 9 "Fundamental Requirements for Military SSB Receiver Design," D. Kahn, Instructor, Fixed Station Equipment, U. S. Army Signal Corps School, Ft. Monmouth, N. J.
- March 16 "Low Noise Preamplifiers," Dr. I. W. Meyer, Associate Director Division 4, M.I.T. Lincoln Lab., Cambridge, Mass.
- March 23 IRE Convention Recess.
- March 30 "Fundamentals of Single Sideband and Some Commercial Practice," S. E. Piller, Group Supervisor, Eldico Electronics Div., Radio Engineering Labs., Inc., New York, N. Y.

URSI-IRE Spring Meeting TO BE HELD IN WASHINGTON

The URSI-IRE spring meeting will be held at the Sheraton Park Hotel and the National Bureau of Standards, Washington, D. C., May 2-5. The IRE Professional Groups on Antennas and Propagation, Circuit Theory, Information Theory, Instrumentation and Microwave Theory and Techniques are cosponsoring the meeting.

The U. S. National Committee, URSI, will hold a business meeting on Monday morning, May 2. A combined technical session for all participants will be held on Monday afternoon, and the Commissions will hold their business sessions on Monday and Tuesday evenings.

The following Commissions are planning to hold one or more technical sessions in addition to their business meetings:

- Commission 1-Radio Measurement Methods and Standards, R. W. Beatty, Chairman.
- Commission 2-Tropospheric Radio Propagation, I. H. Gerks, Chairman.
- Commission 3-Ionospheric Radio Propagation, L. A. Manning, Chairman.
- Commission 4-Radio Noise of Terrestrial Origin, W. Q. Crichlow, Chairman.
- Commission 5-Radio Astronomy, E. F. McClain, Chairman.
- Commission 6-Radio Waves and Circuits. J. I. Bohnert, Chairman.

Calendar of Coming Events and Authors' Deadlines*

(Continued from base 14.4)

- Natl. Conv. on Mil. Elec., Sheraton Park Hotel, Washington, D. C., June 27-29.
- Cong. Intl. Federation of Automatic Control, Moscow. USSR, June 25-July 9.
- Int'l Conf. on Electrical Engrg. Education, Sagamore Conf. Center, Syracuse Univ., Syracuse, N. Y., Jul.
- WESCON, Los Angeles Mem. Sports Arena, Los Angeles. Calif., Aug. 23-26, (DL*: May 1, R. G. Leitner, WESCON Bus. Office, 1435 So. La Cugna Blvd., Los Angeles 35, Calif.)
- URSI 13th Gen. Assembly, Univ. of
- London, London, Eng., Sept. 5-15.
 Joint Automatic Control Conf., M.I.T., Cambridge, Mass., Sept. 7-9.
 Space Electronics and Telemetry Conv.
- and Symp., Shoreham Hotel, Washington, D.C., Sept. 19-22. Industrial Elec. Symp., Cleveland, Ohio,
- Sept. 21-22.
- Sixth Natl. Communications Symp. Hotel Utica and Utica Municipal Aud., Utica, N. Y., Oct. 3-5. (DL*: June 1, B. H. Baldridge, 25 Bolton Rd., New Hartford, N. Y.)
- Natl. Elec. Conf., Hotel Sherman, Chicago, Ill., Oct. 10-12. (DL*: May's 1960 Prof. T. F. Jones, Jr., School of E.E., Purdue Univ., Lafayette, Ind.)
- Symp. on Space Navigation, Deshler-Hilton Hotel, Columbus, Ohio, Oct. 19-21.
- East Coast Conf. on Aero & Nav. Elec., Baltimore, Md., Oct. 24-26.
- 5th Ann. Conf. on Nonlinear Magnetics and Magnetic Amplifiers, Oct. 26-28. (DL*: Mar. 15, D. Katz, Bell Tel. Labs., Inc., Whippany, N. J.)
- Electron Devices Mtg., Hotel Shore-ham, Washington, D. C., Oct. 27-29.
- 13th Ann. Conf. on Elec. Tech. in Med. and Bio., Sheraton Park Hotel, Washington, D. C., Oct. 31, Nov. 1-2.
- Radio Fall Mtg., Hotel Syracuse, Syracuse, N. Y., Oct. 31, Nov. 1-2.
- Mid-Amer. Elec. Conv., Kansas City, Mo., Nov. 14-16.
- 1960 NEREM (Northeast Electronics Res. & Engrg. Mtg.), Boston, Mass., Nov. 15-17.
- PGVC Ann. Mtg., Sheraton Hotel, Philadelphia, Pa., Dec. 1-2.
- Eastern Joint Computer Conf., New Yorker Hotel, New York, N.Y., Dec.

1961

- 7th Natl. Symp. on Reliability and Qual-ity Control, Bellevue-Strafford Hotel, Philadelphia, Pa., Jan. 9-11. (DL*: May 9, 1960, R. E. Kuehn, IBM Corp., Owego, N. Y.
- IRE National Conv., N.Y. Coliseum and Waldorf-Astoria Hotel, New York, N.Y., Mar. 20-23.
- 5th Midwest Symp. on Circuit Theory, Univ. of Illinois, Urbana, May 7-8. (DL*: Oct. 1, M. E. Van Valkenberg, Dept. of E.E. Univ. of Ill., Urbana.)
- Electronic Computer Conf., West Coast, May 9-11.
- WESCON, San Francisco, Calif., Aug. 22-25.

* DL = Deadline for submitting abstracts.

NEC ELECTS 1960 OFFICERS

Dr. Lawrence W. Von Tersch (A'44– M'48–SM'52), head of the electrical engineering department at Michigan State University, has been elected president of the National Electronics Conference for 1960.

Other officers named for the next NEC, which will be held in Chicago at the Hotel Sherman on October 10–12 are:

Executive vice president, Joseph J. Gershon (S'46-M'48-SM'54), DeVry Technical Institute; secretary, James H. Kogen, GPE Controls, Inc.; treasurer, Dr. Harold E. Ellithorn (A'36-M'44-SM'46), University of Notre Dame; assistant Treasurer, Robert J. Parent (M'46), University of Wisconsin.

Dr. Von Tersch, a long-time and active participant in NEC committee work, and formerly vice president of NEC, represents Michigan State University in Conference functions. He is a graduate and received the Ph.D. degree from Iowa State College.

The newly-elected NEC chairman of the board is William O. Swinyard (A'37–M'39– SM'43–F'45), vice-president of Hazeltine Research, Inc. He is a past president of NEC and was a member of the original group responsible for the organization of the Conference.

Committee chairmen elected are:

Arrangements, Benjamin G. Griffith (A'38-VA'39-SM'54), Teletype Corpora-tion; exhibits, John S. Powers (A'44-M'50), Bell and Howell Company; fellowship award, Orville I. Thompson (A'45-M'46-SM'58), DeVry Technical Institute; finance policy, Dr. John D. Ryder (A'29-SM'45-F'52), Michigan State University; housing, Juergen Roedel (S'47-A'50-M'55) Hallicrafters Company; international activities, George E, Anner (A'46-SM'54), University of Illinois; long-range planning, Dr. Christopher E. Barthel, Jr., Armour Research Foundation; NEC party, Stanley I. Cohn (SM'59) Armour Research Foundation; 1959 Proceedings, Clyde H. Hoffman (A'54-M'57-SM'59), University of Notre Dame; 1960 Proceedings, Dr. Thomas L. Butler, Jr. (A'53-M'58), University of Michigan; program, Dr. Thomas F. Jones. Jr. (SM'48), Purdue University; registration, LeRoy W. Murphy, Illinois Bell Telephone Company; student activities, Dr. M. E. Van Valkenburg (S'43-A'45-SM'53), University of Illinois.

The National Electronics Conference, Inc. is a nonprofit organization serving as a national forum for the presentation of authoritative technical papers on electronic research, development and application. More than 10,000 registrants are expected at the 16th annual NEC, which will be held in Chicago, III., in October, 1960.

A few of the activities of NEC include the annual *Fellowship Award* which has a value of \$2500 and is designed to assist worthy undergraduates to further their electronic training; the *Annual Award* which honors an author who participated in the program of the preceding year and presented what was considered the finest paper; and the *Award* of *Merit* which is presented from time to time and honors the author of a particularly influential paper given at any prior Conference. The *Award of Merit* includes a check in the amount of \$750. The last such award was presented to engineer-scientist, Dr. Leon N. Brillouin for his paper on "A Theorem of Larmor and Its Importance for Electrons in Magnetic Fields."

A record attendance and number of exhibitors are expected in 1960. Executive and planning committees are studying new programs in preparation for the anticipated future growth of the Conference. Growth has been stimulated by a management consisting of representatives of the four professional societies and nine educational institutions who participate in the Conference. A full-time staff and general manager conduct Conference business activities at NEC headquarters located in Chicago. Requests for information regarding NEC activities and awards should be directed to the Chicago office.

Sponsors include the AIEE, Illinois Institute of Technology, 1RE, Northwestern University and the University of Illinois. Participants are Michigan, Michigan State, Notre Dame, Purdue, Wayne State and Wisconsin Universities, EIA, and the Society of Motion Picture and Television Engineers.

ELECTRONICS CONFERENCE ISSUES CALL FOR PAPERS

The National Electronics Conference will be held at the Sherman Hotel, Chicago, Ill., on October 10, 11, and 12, 1960. Authors of papers are invited to submit abstracts of 100 to 150 words (for publication in program) and either a 400 to 500 word summary or the completed paper for review. Submit to Professor Thomas F. Jones, Jr., Program Chairman, NEC, School of Electrical Engineering, Purdue University, Lafayette, Ind.

Technical areas typical of those covered at the Conference include: Adaptive Servomechanisms, Antennas and Propagation, Audio, Circuit Theory, Communication Systems, Computers, Information Theory, Instrumentation and Telemetry, Masers, Microminiaturization, Microwaves, Millimeter Waves, Parametric Amplifiers, Plasma Research, Radar and Radio Navigation, Radio Astronomy, Servomechanisms, Signal-Matched Filters, Solid-State Circuits, Solid-State Devices and Materials, Space Electronics, Communications and Navigation, Television, Transistors, and Value Engineering.

AIR FORCE MARS

LISTS MARCH TALKS

The following is the current schedule for the Air Force MARS Eastern Technical Net, which can be heard Sundays from 2 to 4 P.M. EST, at 3295, 7540, and 15,715 kc.

- March 6 "The IRE National Convention," G. Bailey. Chairman of the Convention.
- March 13 "Sonar Transducers," J. Campani, Chief Engr., Telephonics Corp.
- March 20 "Energy Systems for Space Operations," Dr. K. F. Rubert, Scientist, NASA, Langley Field, Va.
- March 27 "The Jet Pilot," Flight Lt. H. Clark, RCAF.
- April 3 "Tool Design," F. Yesmant, Engr., Riverside Plastics Corp.

DENVER RESEARCH INSTITUTE ANNOUNCES PAPER DEADLINE

The Denver Research Institute of the University of Denver will hold its 7th Annual Symposium on Computers and Data Processing at the Stanley Hotel in Estes Park, Colo., on July 28 and 29, 1960. The continuing theme of this series of meetings has been the advanced treatment of basic problems in computer technology. Papers will be presented in the fields of Components and Devices, Logic Design, and Philosophy of Computer Design.

Although it is anticipated that the program will be comprised largely of invited papers, a limited number will be selected from papers submitted without invitation. Authors wishing to submit papers may send abstracts of approximately 150 words, no later than April 1, 1960, to: W. II. Eichelberger, Denver Research Institute, University of Denver, Denver 10, Colo.

Electronics Marketing Business Letter Service Established in New York

Shepherd's *Electronics Marketing*, a monthly business letter service, was introduced in January, 1960. This publication, plus supplements, presents marketing ideas and information exclusively for the electronics industries. It is published by Shepherd Associates, 130 West 42 St., New York, N. Y., headed by Michael R. Shepherd, publisher, and is edited by Sidney Feldman (A'58), former marketing editor of *Electronic Week*, associate editor of *Forbes Magazine*, and contributor to publications including *The New York Times*, Barron's and others.

The new publication covers electronics markets, as well as marketing management, services and personnel.

WILLOW RUN LABS WILL HOLD 6th Annual Radar Symposium

Since 1955, annual radar symposiums have been held at The University of Michigan, Ann Arbor, in recognition of the fact that the dissemination of information among scientists and engineers working in the field of radar is necessary to improve radar techniques, devices, and applications. Continuing the series, the Radar Laboratory of the Willow Run Laboratories at The University of Michigan will conduct the Sixth Annual Radar Symposium at Ann Arbor on June 1-3, 1960, with the support of Project MICHIGAN and under sponsorship of the Army, Navy, and Air Force. Project MICHI-GAN, which engages in research and development for the U.S. Army Combat Surveillance Agency, is carried on by the Willow Run Laboratories under Department of the Army Contract DA-36-039 SC-78801, administered by the U.S. Army Signal Corps.

The tentative program consists of single general sessions each morning and multiple specialized sessions and panel discussions each afternoon. Papers will pertain to the general field of radar, with particular emphasis on these major areas: Components and Techniques, Propagation Phenomena, Engineering Applications, and New Data and Their Organization.

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Life -- up to 500 hours guaranteed -- over 3000 hours reported Frequency stability -- less than 2 Mc drift per 100 hours (C band) Power stability -- drop of less than 1 db for constant voltage input Duty cycle stability -- less than 3 Mc frequency shift for a change in duty cycle of 0.00005 to 0.002 (C band) Vibration -- less than 2.5 Mc frequency shift from 55 to 2000 cps Shock -- withstands 100 g's of 6 millisecond duration Lightweight -- 7 to 10 oz. Miniaturized Tunable over a broad band





Band	Tube Type	Fixed Freq. or Tunable	Frequency Range Mc	Min. Peak Power Watts	Outptut Mates With	BETHIC
C	BL-212	Tunable	5400-5900	100	UG699/U	New short form cathlog available. Send for your copy today,
C	BL-243	Tunable	5400-5900	200	UG699/U	
C	BL-242	Tunable	5400-5900	400	N	
C	BLN8-022	Tunable	5400-5900	500	TNC	BOMAC Inhermiteries inc
C	BLM-026	Tunable	5400-4900	500	TNC	ANNUM - THE CALL - CALCOLOGICAL HIGH
C	BLM-020	Tunable	5400-5900	700	TNC	
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PROCEEDINGS OF THE IRE Mart

March, 1960

NAB SELECTS T.A.M. CRAVEN FOR ENGINEERING AWARD

The National Association of Broadcasters has announced that it will present its second annual Engineering Achievement Award to Federal Communications Commissioner T. A. M. Craven (F'29) for his "long and distinguished career" during which he has "measurably advanced the technical state of broadcasting through his long and vigorous leadership in engineering activities.

The award will be presented to him on April 6 at a joint engineering-management luncheon of the NAB's Annual Convention in Chicago.

Commissioner Craven was selected by an Awards Subcommittee of the NAB's Broadcast Engineering Conference Committee.

He has just completed an assignment as chairman of the United States Delegation, Administrative Radio Conference, International Telecommunications Union in Geneva, Switzerland. His career includes participation on most of the principal international conventions on communications. He also was a pioneer in adopting directional antennas to facilitate the growth of broadcasting.

He began a seven-year term on the Federal Communications Commission on June 30, 1956, following appointment by President Eisenhower. He previously had served on the Commission during the Roosevelt administration and was the first commissioner to receive nonconsecutive appointments. His experience in Government regulation of electrical communication began in 1928 when, for about two years, he served on the staff of the Federal Radio Commission in charge of nonbroadcast engineering matters. He first joined the FCC as its Chief Engineer in 1935, which post he held until he became a Commissioner.

Born January 31, 1893 in Philadelphia, Pa., he was graduated from the U.S. Naval Academy in the class of 1913. During his naval career he specialized in radio communication. He served afloat as radio officer of various battleships and as fleet radio officer, first of the U.S. Asiatic Fleet and then the combined U.S. Fleet. He was responsible for the first modernization of the fleet radio communications system and received a special commendation for his World War I development of a means of transmitting orders to ships in submarine-infested waters without disclosing vessel positions.

During his naval and civilian Government career, he has been a member of and at one time chairman of the Interdepartment Radio Advisory Committee. Since 1919, he has been either a technical advisor, a member, or chairman of U.S. delegations to various international radio conferences including

the Allied Radio Consulting Committee, Communications Conference of Allied and Associated Powers, Provisional Inter-Allied Radio Technical Consulting Committee, International Radio Conference (for which service he was commended by the Secretary of the Navy), First International Radio Consulting Committee, United States-Canadian Radio Conference, International Radio Conference, Inter-American Radio Conference, and North American Regional Broadcasting Conference.

In 1930 he resigned his commission as a Lieutenant Commander in the regular Navy and from then, until 1944, served as a Commander in the U.S. Naval Reserve.

From 1930 to 1935 he was vice president in charge of technical matters of several broadcasting companies. During that period he also engaged in private practice as a consulting radio engineer. From 1944 to 1949 he was vice president in charge of technical matters for the Cowles Broadcasting Co. At the same time he was a member of the board of directors of the National Association of Broadcasters. In 1949 he became a member of the firm of Craven, Lohnes and Culver, Washington, D. C. He is a past president of the Association of Federal Communications Commission Consulting Engineers.

CINCINNATI CONFERENCE To Be Held in April

The Spring Technical Conference of the Cincinnati Section of the IRE will be held on Tuesday and Wednesday, April 12 and 13, 1960, at the Hotel Alms in Cincinnati, Ohio.

Under the direction of John Ebbeler, the committee has decided to present the Fourteenth Annual Conference in conjunction with the American Rocket Society. The conference themes are electronic data processing and space technology.

The committee members for the conference are J. R. Ebbeler, Conference Chairman; A. B. Ashman, Registration Chairman; R. P. Schlemmer, Advertising Chairman; S. W. Stuhlbarg, Papers Chairman; C. F. Winder, Banquet Chairman; R. H. Lehman, Arrangements Chairman; and V. Scott, Publicity Chairman.

The sessions will be:

- I. Venus, Target for Tonight
- II. Re-entry
- III. Bio-Electronics
- IV. Objectives of the IGY
- V. Inertial Guidance
- VI. Can Machines Outthink People?
- VII. Electronic Data Processing Comes of Age
- VIII. Why Will Tunnel Diodes Revolutionize Electronics?

OBITUARY

James W. McRae (A'37-F'47), Vice President of AT&T Co. and 1953 President of the IRE, died recently at the age of 49. Born on October



J. W. MCRAE

25, 1910 in Vancouver, B. C., Canada, Dr. McRae received the B.S.E.E. degree in 1933 from the University of British Columbia. He received the M.S. and Ph.D. degrees from California Institute of Technology in 1934 and 1937, respectively.

Early in 1937 he

had joined the Bell Telephone Laboratories, where he engaged in research on transoceanic radio transmitters. After working on microwave research, he worked on military projects, including a special microwave oscillator for the NDRC and early association with several microwave radar projects. After serving in the U.S. Army Signal Corps in World War II, he returned to Bell Labs. to become Director of Radio Projects and Television Research in 1946. In 1947 he became Director of Electronic and Television Research, which made him responsible for electron dynamics research. He was appointed Director of Apparatus Development I in 1949, and later that year he became Director of Transmission Development. In 1951 he was elected a Vice President of the Laboratories, in charge of the Systems Organization, responsible for switching and transmission development and systems engineering. In 1953 he was elected a Vice-President to Western Electric Co. and President of Sandia Corp. In 1958 he was elected Vice President of AT&T, responsible for the defense activities of the Bell System.

In 1958 he was appointed to the General Advisory Committee to the Atomic Energy Commission, and in 1959 he was named Chairman of the Army Scientific Advisory Panel.

Dr. McRae was a member of the Board of Editors of the IRE from 1946 to 1949. He was chairman of the New York Section in 1949, Chairman of the Ad Hoc Committee on Technical Groups, and Chairman of a Committee on Section vs Subsection Problems in 1952. He was a member of the Board of Directors and the Executive Committee from 1949 to 1955. He also served on a number of other Institute committees.

Dr. McRae received honorary mention from Eta Kappa Nu as an outstanding young electrical engineer in 1943, and he received the Legion of Merit for his work in radar during World War II. He was a member of AIEE and Sigma Xi.

1960 Nuclear Congress

NEW YORK COLISEUM, NEW YORK, N. Y., APRIL 4-7, 1960

A partial program of sessions planned for the 1960 Nuclear Congress, to be held in the New York Coliseum, April 4-7, has been announced by Dr. Clarke Williams, chairman of the Nuclear Congress.

The Congress, a gathering of representatives from all areas of the nuclear field, is sponsored by 28 leading engineering, scientific, management, and technical organizations. It consists of the 6th Nuclear Engineering and Science Conference, the 8th NICB Atomic Energy in Industry Conference, and the 6th International Atomic Exposition.

The Exposition, which was established in 1954, will include at least 130 exhibits of the manifold products and services available for the peaceful use of atomic energy. More than 1000 requests for information regarding participation in the exhibit have been received from firms all over the world.

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The theme of the meeting, according to E. B. Gunvou, program committee chairman of the Nuclear Congress, will be "What Will the Future Development of Nuclear Energy Demand from Engineers?" This question will be approached through a series of reports, papers, and discussions on a wide variety of subjects related to the peaceful use of atomic energy.

Environment Session I

Chairman: J. C. Callahan, Morris Knowles, Inc.

"Current Cost and Construction Experience at Yankee Reactor Site," C. T. Chave, Stone & Webster, Boston, Mass.

"Pressure Suppression," F. F. Mautz, Pacific Gas & Electric Co., San Francisco, Calif.

"Pressure Liquidation," A. F. Kolflat, Sargent & Lundy, Chicago, Ill.

"Review of Shippingport Operations-Site Factors," J. E. Gray, Duquesne Light Co., Pittsburgh, Pa.

"A Progress Report on the N.S. Savannah," P. P. Eddy and K. W. Hess, Nuclear Projects Office, Washington, D. C.

Environment Session II

Chairman: M. Eisenbud, New York Univ., New York, N. Y.

"Local Problems in Regulation," H. Blatz, New York City Health Dept., New York, N. Y

"State Food Problems," L. Menzer, Hartford City Health Dept., Hartford, Conn.

"Needs for Uniformity in Laws. "International Problems in Radiation,"

Representative of the U. N.

Environment Session III

Chairmen: E. S. Cole, Pitometer Assocs., and R. E. Fuhrman, Fed. of Sewage & Indust. Wastes Assoc.

"Environmental Radioactivity in Large Area Surrounding Nuclear Electric Plant," J. V. Nehemias, Radiological Health Surveys of National Sanitation Found., Univ. of Michigan, Ann Arbor.

"Strontium 90 in Surface Waters," C. P. Straub, L. R. Setter, P. F. Hallbach, Dept. of Health, Education and Welfare, Washington, D. C., and R. A. Taft Engrg. Center, Cincinnati, Ohio.

"Use of Tritium as a Tracer in Evaluating Waste Discharges," W. J. Kaufman and R. M. Hours, Univ. of California, Berkeley.

"Hanford Reactor Effluent Contribution to Environmental Radiation Dose," R. L. Junkins, General Electric Co., Richland, Wash.

Materials and Components Session I

Chairman: B. W. Dunnington, Battelle Memorial Inst., Columbus, Ohio.

"Effect of Irradiation on Tensile Properties of High-Temperature Steel and NonFerrous Allovs," S. Bartz, Phillips Petroleum

"Experience in Use of Organic Coolant in Nuclear Reactors," H. Perlman, Atomics International, Canoga Park, Calif.

"Problems in Contamination of Water-Cooled Reactors Using Stainless Steel Components," D. Foley, Alcoa, Inc., Schenectady, N. Y.

Materials and Components Session II

"Ignition and Detonation of Uranium in Boron Trifluoride Solutions," R. Johnson, F. Horn and G. Strickland, Brookhaven National Lab., Upton, N. Y.

"Uranium-Thorium Reprocessing; Cost Outlook for Commercial Operation," R. J. Klotzbach, Union Carbide Nuclear Co., New York. N. Y.

"The Plutonium Cycle," J. J. Cadwell, Hanford Atomic Products Operation, General Electric Co., Richland, Wash.

"Nuclear Fuel in West Germany," A. Boettcher, Degussa, Frankfort, Germany.

"Low Impurity Core Material," G. Pancer and J. L. Zegger, Alco Products, Schenectadv, N. Y.

Reactor Session I

U. S. Power Reactors

Chairman: S. Baron or B. Noyes, Burns & Roc. Inc., New York, N. Y.

"The Army Nuclear Power Program," Col. D. G. Williams, Chief, Arniy Nuclear Power Program.

"BUDÖCKS Role in the Development of Nuclear Power for the Navy's Shore Establishments," Comm. W. J. Christensen, CEC, U. S. Navy.

Reactor Session II

Research Reactors and Radiation Facilities

Chairman: H. Neal, American Machine & Foundry.

"BMI Reactor Operations and Interest-ing Experiments," J. Chastane.

"How to Live with Experimenters Operating Problems with New Reactors," J. Cox, Oak Ridge National Lab., Oak Ridge, Tenn

"McMaster University Reactor Experimental Operations," Dr. Thod or W. Fleming, McMaster Univ., Hamilton, Ont.

"BNL Medical Reactor with a Survey of Other Reactor Operating Difficulties and Remedies," R. Burrell, Brookhaven National Lab., Upton, N. Y.

Reactor Session III

Progress in Reactor Instrumentation

Chairman: H. A. Lamonds, North Carolina State College, Raleigh.

"Application of Control and Instrumen-W. Lapinski, Argonne National tation," Lab.

"Application and Effect of Control and Instrumentation," C. F. Obermesser, West-inghouse Electric Corp.

"C. & I Equipment-In-Core," H. M.

Ogle, General Electric Co., San Jose, Calif. "C & I Equipment—External," E. P. Epler, Oak Ridge National Lab., Oak Ridge, Tenn.

"Significance and Need of C & I," H. E. Vann, Atomic Energy Commission.

Reactor Session IV

Advanced Reactor and Fuel Cycles

Chairman: L. J. Everett, Philadelphia Electric Co., Philadelphia, Pa.

"Advanced Water Cooled Reactor," representative of Westinghouse.

"Army Gas Cooled Reactor Systems Program," M. A. Rosen and Capt. G. A. Bicher.

Nuclear Propelled Aircraft Session I

"Some Practical Methods for Fabricating Shields for Nuclear Powered Aircraft, W. Q. Hullings and J. L. McDaniel, Covair, Fort Worth, Texas.

"Nuclear Propulsion for Lifting (Large) Space-Stations into Orbit," Dr. R. A. Mayer, Norair Div. of Northrop Corp., Hawthorne, Calif.

"Nuclear Ramjet Developments," T. C. Merkel, Univ. of California, Radiation Lab., Livermore, Calif.

"The Effect of Gamma Radiation Upon an Electro-Hydraulic Servo System," R. N. Miller and W. C. Bennett, Lockheed Aircraft Corp., Marietta, Ga.

Aircraft Nuclear Propulsion Session II

"The Nuclear Propelled Airship," L. Jurich, Goodyear Aircraft Corp., Akron, Ohio.

"Potential of Nuclear Powered Aircraft for Commercial Cargo," J. F. Brady, Jr., Covair, San Diego, Calif.

"Three Approaches to Achieving Reliability for Nuclear Powered Navy Aircraft," L. Credit, Nuclear Div., The Martin Co., Baltimore, Md.

Nuclear Standards Session

Chairman: F. L. LaQue, International Nickel Co., New York, N.Y.

N. L. Mochel, Westinghouse Electric Corp., Lester Station, Philadelphia, Pa., will speak for the American Society for Testing Materials, of which he is a past president.

Dr. C. R. McCullough, Atomic Energy Commission, Washington, D. C., who is chairman of the American Nuclear Society Standards Committee, will speak for that group.

E. Bailey, Commonwealth Edison Co., Chicago, Ill., will speak for the American Society of Mechanical Engineers' Special Committee on Nuclear Power of the Boiler and Pressure Vessel Committee.

J. A. Klapper, Ebasco Service Inc., New York, N. Y., will speak for American Standards Association. He is chairman of the Materials Task Force, Nuclear Advancement Committee, of the Standards Association.

Isotope Application Session

"Industrial Radioisotope Applications Development Sponsored by AEC," P. G. Acbersold, Office of Isotopes Development, United States Atomic Energy Commission, Washington, D. C.

"Isotopes in Petroleum Production," M. Williams, Humble Oil & Refining Co., Houston, Texas.

"Industrial Activities in Applications of Radioisotopes," P. Kruger, Nuclear Science

& Engineering Corp., Pittsburgh, Pa. "Isotopic Power," J. G. Morse, Isotopic Power Dept. The Martin Co., Baltimore, Md.

"Preparation, Properties and Applications of Radioactive Clathrates," D. J. Chleck, Tracerlab Inc., Waltham, Mass.

March, 1960



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- Antennas & Propagation (G-3)—A. Dorne, Dorne and Margolin, Westbury, L. L., N. Y.; S. A. Bowhill, Pennsylvania State Univ., University Park, Pa.
- Audio (G-1)—Prof. A. B. Bereskin, E.E. Dept., Univ. of Cincinnati, Cincinnati 21, Ohio; M. Camras, Armour Res. Found. Tech. Ctr., Chicago 16, Ill.
- Automatic Control (G-23)—J. E. Ward, Servomechanisms Lab., M.I.T., Cambridge 39, Mass.; G. S. Axelby, Westinghouse Air Arm Div., Friendship Airport, Baltimore 3, Md.
- Broadcast & Television Receivers (G-8)— R. R. Thalner, Sylvania Home Electronics, 700 Ellicott St., Batavia, N. Y.; C. W. Sall, RCA, Bldg, 13–4, Camden, N. J.
- Broadcasting (G-2)—G. E. Hagerty, Westinghouse Elec. Corp., 122 E. 42 St., Suite 2100, N. Y. 17, N. Y.; W. L. Hughes, E.E. Dept., Iowa State College, Ames, Iowa.
- Circuit Theory (G-4)—S. Darlington, Bell Telephone Labs., Murray Hill, N. J.; W. Bennett, Bell Telephone Labs., Murray Hill, N. J.
- Communications Systems (G-19)—J. E. Schlaikjer, IT&T Co., 67 Broad St., N. Y. 4, N. Y.; M. R. Donaldson, Electronic Comm. Inc., St. Petersburg, Fla.
- Component Parts (G-21)—J. J. Drvostep, Sperry Gyroscope Co., Mail Station 1A 36, Great Neck, L. I., N. Y.; G. Shapiro, Engineering Electronics Sec. Div. 1.6, NBS, Connecticut Ave. and Van Ness St., Washington, D. C.

* Name listed are Group Chairmen and Transactions Editors,

- Education (G-25)—J. G. Truxal, Head, Dept. of E.E., Polytechnic Inst. of Brooklyn, Brooklyn, N. Y.; W. R. LePage, Dept. of E.E., Syracuse Univ., Syracuse, N.Y.
- Electron Devices (G-15)—W. M. Webster, Semi-Conductor Div., RCA, Somerville, N. J.; E. L. Steele, Hughes Prods., Inc., International Airport Station, Los Angeles 45, Calif.
- Electronic Computers (G-16)—R. O. Endres, Rese Engrg. Co., Philadelphia, Pa.; H. E. Tompkins, Moore School of E.E., Univ. of Pennsylvania, Philadelphia, Pa.
- Engineering Management (G-14)—II. M. O'Bryan, Sylvania Elec. Prods., 730 3rd Ave., N. Y. 17, N. Y.; A. H. Rubenstein, Northwestern Univ., Evanston, Ill.
- Engineering Writing and Speech (G-16)— T. T. Patterson, Jr., RCA, Bldg. 13–2, Camden, N. J.; J. Kinn, Electronics, 330 W. 42 St., N. Y., N. Y.
- Human Factors in Electronics (G-28)— C. M. Jansky, Royal McBee Corp., Portchester, N. Y.; J. I. Elkind, Bolt, Beranek and Newman, Cambridge, Mass.
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- Information Theory (G-12)—P. Elias, M.I.T., Rm. 26–347, Cambridge 39, Mass.; G. A. Deschamps, E.E. Dept., Univ. of Illinois, Urbana, Ill.
- Instrumentation (G-9)—C. W. Little, C-Stellerator Assoc., Box 451, Princeton, N. J.; G. B. Hoadley, Dept. of E.E., North Carolina State College, Raleigh, N. C.
- Medical Electronics (G-18)—W. E. Tolles, Airborne Instruments Lab., 160 Old Country Rd., Mineola, L. I., N. Y.; L. B. Lusted, Univ. of Rochester Medical

School, Strong Memorial Hosp., Rochester 20, N. Y.

- Microwave Theory and Techniques (G-17) —A. A. Oliner, Microwave Res. Inst., 55 Johnson St., Brooklyn 1, N. Y.; D. D. King, Electronic Comm., Inc., 1830 York Rd., Timonium, Md.
- Military Electronics (G-24)—H. Randall, 1208 Seaton Lane, Falls Church, Va.; D. R. Rhodes, Radiation Lab., Instrument Div., Orlando, Fla.
- Nuclear Science (G-5)—A. B. Van Rennes, United Res. Inc., Tech. Div., 128 Alewife Brook Pkwy., Cambridge, Mass.; R. F. Shea, Dig Power Plant Engrg., Knolls Atomic Power Lab., General Electric Co., Schenectady, N. Y.
- Production Techniques (G-22)—L. M. Ewing, General Electric Co., HMEED CSP-3, Syracuse 1, N. Y.; A. R. Gray, Rte. #1, Box 940, Orlando Vineland Rd., Wintergarden, Fla.
- Radio Frequency Interference (G-27)— J. P. McNaul, U. S. Signal Corps., Hdqrs. Ft. Monmouth, N. J.; P. O. Schreiber, Technical Wire Prods., Springfield, N. J.
- Reliability and Quality Control (G-7)— P. K. McElroy, General Radio Co., 22 Baker Ave., West Concord, Mass.; E. J. Breiding, IBM Corp., Kingston, N. Y.
- Space Electronics and Telemetry (G-10)— C. H. Hoeppner, Radiation, Inc., Melbourne, Fla.
- Ultrasonics Engineering (G-20)—W. Roth, Roth Lab., 1240 Main St., Hartford 3, Conn.; O. Mattiat, Aerophysics Dev. Corp., P.O. Box 689, Santa Barbara, Calif.
- Vehicular Communications (G-6)—A. A. MacDonald, Motorola, Inc., 4545 Augusta Blvd., Chicago 51, Ill.; R. P. Gifford, General Electric Co., Syracuse, N. Y.

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- Albuquerque-Los Alamos (7)—R. C. Spence, 3020 W. Sandia Dr., Sandia Base, Albuquerque, N. Mex.; E. C. Davis, Neely Enterprises, 107 Washington St. S. E., Albuquerque, N. Mex.
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- Cedar Rapids (5)—D. O. McCoy 2315 Blake Blvd., Cedar Rapids, Iowa; G. M. Hodgin, R 3, Marion, Iowa.
- Central Florida (3)—W. S. Hines, 1320 Indian River Dr., Eau Gallie, Fla.; C. E. Mattox, 209 Beverly Re., Cocoa, Fla.
- Central Pennsylvania (4)—C. R. Ammerman, 125 Grandview Rd., State College, Pa.; W. J. Ross, 105 Elec. Engrg. Dept., University Park, Pa.
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World Radio History

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Lt. Colonel Melvin N. Abramovich $(S'36 - \Lambda'46 - M'47 - SM'50)$. USAF, has been recently appointed Chief of the Wash-

ington, D. C. Regional Office of the Air Research and Development Command. In this position he will direct the technical liaison activities of that office for Head-ARDC quarters and its agencies. The Washington Regional Office is part of a nation-



M. N. Abramovich

wide system of Regional and Liaison Offices established by ARDC to facilitate technical coordination and liaison between Air Force research and developmental agencies and other government, industry, public and private laboratories and institutions engaged in research, engineering, testing and evaluation of technical equipments and systems. Coverage includes work in all fields of science and engineering and in the various technical areas in which the AF research and development program is divided. Three additional Regional Offices are located in New York City, Chicago, and Los Angeles. Each of these offices, in turn, has a number of satellite Liaison Offices located at other geographical centers of engineering activities. The services of all these offices are available to industry to aid them in their dealings with the Air Research and Development Command complex

Lt. Colonel Abramovich is a graduate of the Institute of Technology of the University of Minnesota and has received additional training at Massachusetts Institute of Technology and George Washington University. He also attended the Army's Signal Corps School at Fort Monmouth, N. J. He later joined the Curtiss-Wright Corp. In 1940 he was called to active duty and served in various electronic and communications assignments including the Communications Department of the Armored Force School, at Fort Knox, R&D Division of the Office of the Chief Signal Officer, Watson Laboratories (now Rome Air Development Center) and AAF Liaison Office of the Navy's Bureau of Ships, and Headquarters USAF. While overseas during World War II, he served as Deputy Chief Radar Officer of the Mediterranean Allied Coastal Air Force. Upon returning to civilian status, he joined Cambridge Electronics Corporation of Baltimore. In 1947 he was again recalled to active duty to become the technical liaison officer for the Engineering Division of the Air Materiel Command with duty station at the Naval Research Laboratory. He later served in the Electronics Directorate at Headquarters ARDC. In 1954 he was selected to become a member of the staff of the Director of Electronics for the Assistant Secretary of Defense (Research & Development). In this latter position, he was instrumental in establishing the Department of Defense Electronics Test Equipment Coordination Group and served as its first Executive Secretary and later as its Chairman. Since 1957 and until his appointment as Chief, Lt. Colonel Abramovich has been Electronics Staff Officer and Deputy Chief of the ARDC Regional Office, Washington, D. C.

He has been AF member of the Panel on Tubes, Panel on Component Parts, Panel on Test Equipment, and Advisory Group on Reliability of Electronic Equipment of the former Research and Development Board of the Department of Defense. He was also member of Joint Test Equipment Subpanel of the Joint Communication—Electronics Committee, Joint Chiefs of Staff.

Lt. Colonel Abramovich is a member of the American Association for the Advancement of Science, the Institute of Navigation, and was a Charter Member of the Engineers Club of Washington, He is a Registered Professional Engineer of the District of Columbia.

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Professor S. V. C. Aiya (SM'43) is President of the Institution of Telecommunication Engineers, India, for 1959-60.

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(Continued on page 32A)

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2N1015 2N1015A 2N1015B 2N1015C 2N1015D	30 60 100 150 200	10 @ I _c =2 amp	.75 ohms @ I _c =2 amp I _a =300 ma	7.5	150°C	.7°C/W	
2N1016 2N1016A 2N1016B 2N1016C 2N1016D	30 60 100 150 200	10 @I _c =5 amp	.50 ohms @ I _c =5 amp I _e =750 ma	7.5	150°C	.7°C/W	
*TRUE voltage rating (The transistors can be operated continuously at the Vcz listed for each rating.)							

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PROCEEDINGS OF THE IRE

March, 1960



HOSKINS Chromel-R



A modified 80-20 type nickel-chromium alloy possessing optimum uniformity of all physical properties required for close tolerance electronic control applications. Possesses electrical resistivity of

control applications. Possesses electrical resistivity of 800 ohms/cmf at 20°C. and a low temperature coefficient controlled within 0 ± 10 ppm/°C. Performance characteristics include remarkably low noise level plus exceptional linearity and stability from -65° to $+150^{\circ}$ C.

HOSKINS Alloy 815-R



A lower density, higher resistivity iron-chromiumaluminum composition that gives you 14% more ohms per pound than nickel-chromium resistor alloys. It possesses high strength, good ductility, excellent resistance

alloys. It possesses high strength, good ductility, excellent resistance to wear and corrosion. Specific resistance is 815 ohms/cmf at 20°C. and temperature coefficient is inherently controlled within 0 ± 10 ppm/°C. over the range from -65° to $+150^{\circ}$ C.

If you make potentiometers or precision wire wound resistors, these alloys are *right* for you—*right* for your customers, too. Complete technical data—the most comprehensive ever offered—are available upon request, as are sample spools of both alloys taken from current production material. Send for them today!

HOSKINS MANUFACTURING COMPANY

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(Continued from page 30A)

tion of telecommunication and electronic equipment is arranged. It has several local centers in different parts of India which arrange for periodic paper meetings and discussions.

Professor Aiya, its new President, had a distinguished university career at Bombay and Cambridge (England) and has held positions in academic institutions and universities at Bombay, Poona and Ahmedabad and is now professor and head of the department of Electrical Communication Engineering at the Indian Institute of Science, Bangalore, India's premier research institute. He has from time to time served on several important committees of Government dealing with technical education and research, etc., and assisted the Army authorities in radio work during World War H. He is, at present, a member of the Radio Research Committee of the Indian Council of Scientific Research, the working group on Technical Education of the Indian Planning Commission, the Executive Committee of the Indian Electronics Engineering Research Institute, the Defence Electronics Research Committee, the Expert Committee on Engineering Education of the Institution of Engineers, India, and other committees. He is India's representative on Commission IV of URSI. He is a member of the IEE, London, and the Institution of Engineers, Indla.

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Tore N. Anderson (S'47-A'49-SM'55) has joined FNR, Inc. of Woodside, N. Y. to fill the newly-created position of assist-

ant to Henry Feldmann, the President.

Mr. Anderson comes to FXR from Airtron, Inc., a division of Litton Industries. He joined Airtron in 1948, was appointed Chief Engineer in 1951, elected a Director and Vice-President in 1953.



T. N. Anderson

and became Director of Engineering in 1956.

He has been Chairman of the Waveguide Connector Standardization Subcommittee of the Electronics Industry Association (EIA) since 1954. He recently participated in meetings held in Uhn, Germany, for the preparation of an international standard of waveguide and wave-

(Continued on page 34.4)



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March, 1960



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*Life tests have proved that El-Menco Mylar-Paper Dipped Capacitors --- tested at 100°C with rated voltage applied --have yielded a failure rate of only 1 per 716,800 unit-hours for 1 MFD. Since the number of unit-hours of these capacitors is inversely proportional to the capacitance, 0.1 MFD El-Menco Mylar-Paper Dipped Capacitors will yield ONLY 1 FAILURE IN 7,168,000 UNIT-HOURS.

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Five case sizes in working voltages and ranges:

200 WVDC	.018 to .5 MFD
400 WVDC	.0082 to .33 MFD
600 WVDC	.0018 to .25 MFD
1000 WVDC	.001 to .1 MMF
1600 WVDC	.001 to .05 MFD

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THESE CAPACITORS WILL EXCEED ALL THE ELEC-TRICAL REQUIREMENTS OF E.I.A. SPECIFICATION RS-164 AND MILITARY SPECIFICATIONS #MIL-C-91A AND MIL-C-25A.

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INSULATION Durez phenolic resin impregnated.

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For .05MFD or less, 100,000 megohms minimum. Greater than .05 MFD, 5000 megohm-micrafarads.

For .05MFD or less, 1400 megohms minimum. Greater than .05MFD, 70 megohm-microfarads.

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(Continued from page 32.4)

guide connectors. In 1959, he was appointed a consultant to the Department of Defense, Advisory Group on Electron Parts, Working Group on Transmission Lines.

Mr. Anderson was Chairman of the Northern New Jersey Section of the IRE and organized the Chapter's Professional Group on Microwave Theory and Techniques. He is now on the Administrative Committee and Editorial Board of the PGMTT of IRE. He is also a member of the American Physical Society and the American Institute of Electrical Engineering. He was elected to Tau Beta Pi and Mu Alpha Omicron. He received the B.S. degree from Cooper Union.

•

David R. Baldauf (A'52–M'57) has been appointed development engineer in Systems Analysis at the Owego facility of HBM's Federal Systems Division.

He joined IBM in 1952 as a design engineer in Radar Display Development, and was named associate engineer in 1954, project engineer, manager of AN/ASQ-38 Electronic Development in 1955, and advisory engineer, Systems Coordination and Planning in 1958.

Originally from Massillon, Ohio, Mr. Baldauf earned a B.S. degree in Electrical Engineering from Purdue University in 1952. He has taken graduate work through the HBM-Syracuse University program. He served with the U. S. Navy during World War H as an electronics technician.

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The election of **Alfred S. Backus** (A/52– SM/59) as Vice President, Operations, of the Mycalex Corporation of America and its affiliated com-

ns annated companies, Mycalex Electronics Corporation, Mycalex Tube Socket Corporation and the Synthetic Mica Company, has been announced by Jerome Taishoff, Mycalex President.

Mr. Backus has been serving as Acting General



A. S. BACKUS

Manager of the Mycalex Corporation of America and its affiliated companies for the past year and has been Works Manager since 1952. He joined the Mycalex Corporation of America in 1944. He served originally as Plant Superintendent, becoming Plant Manager and then Works Manager. Prior to joining Mycalex, he had been employed by the General Electric Company, Chemical Division, in their Pittsfield and Taunton, Mass. plants.

Mr. Backus is a member of the Society of Plastics Engineers and the Society of Metal Production.

Continued on page (6.1)

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PROCEEDINGS OF THE IRE March, 1960



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The best way to tell is to know the manufacturer. If you're not already a customer we'd like to number you among those who know from experience that they have transformer reliability inside. when it says ADC on the outside. Over 15,000 custom transformer designs in nearly 25 years have proven beyond a doubt that long life just plain runs in the ADC family.

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range interceptor missile He previously had been project engineer for four years at Aircraft Armaments.

Inc., where he was in charge of systems development and design of equipment associated with the guidance of the TALOS and TERRIER missiles. He began his professional career in 1947 as a member of the research and development department at Bendix Radio, Towson, Maryland, where he was engaged in microwave work and antenna design.

In twelve years with American industry, Dr. Castruccio has to his credit some 200 disclosures and about 20 patents pending or issued.

A member and Vice President of the Maryland Chapter of the American Rocket Society, he received the annual award of the Maryland Chapter of the American Rocket Society in 1958 for his "Outstanding Accomplishments in the Field of Rocketry.

Dr. Castruccio became technical director of Aeronca's Aerospace Division in mid-1959. Creation of the division signaled the formal entry of Aeronca into the space field.

.....

Hazeltine Research Corporation has announced the appointment of William F. (S'33-A'36-VA'39-SM'51-F'54), Bailey Richard J. Farber (S'45-A'47-SM'54) and Donald Richman (S'42-A'45-SM'52) as

(Continued on page 38.4)



P. A. CASTRUCCIO

neering work in astronautics and space technology and his accomplishments in the design and development of a satellite and space ship stabilization system and a space guidance system.

IRE People

(Continued from page 34.4) Dr. Peter A Castruccio (SM'57), 34, technical director of the Aerospace Divi-

sion of Aeronca Manufacturing Corpora-

been

pio-

tion in Baltimore,

has

named one of the

outstanding young men of 1959 by the United States Junior Chamber of

Recognition as

one of 1959's out-

standing young men is based on Dr.

Castruccio's

Md.,

Commerce.

Dr. Castruccio's experience covers a broad range of technical activity, including radar systems, navigational systems, circuits, instrumentation, servomechanisms, computers, automation, space technology missile and space system analysis and design, and mechanical engineering.

In 1958, Dr. Castruccio was Director

of the Astronautics Institute charged with the technical planning, administration and coordination of space programs for Westinghouse Electric Company, From 1955 to 1957, he was head of the Preliminary Design Section, Air Arm Division of Westinghouse, engaged primarily in systems analysis and preliminary design of the terminal guidance system of the BOMARC long-

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Minneapalis 7, Minnesota

March, 1960

RELIABLE SILICON TRANSISTOR SWITCHING



9 COMPONENTS REPLACED BY 4



HOW? – By using Fairchild's 2N1252 or 2N1253 **lowstorage** silicon mesa transistors. The guaranteed low storage characteristic permits a simple saturating circuit to achieve switching speeds that previously required complex non-saturating circuits.

WHY? – Improved reliability and reduced cost – one semiconductor instead of five and fewer soldered connections. Power dissipation is only 1/3rd to 1/5th as great, making possible much higher component densities in packaging. Cost and reliability are improved all the way from development through volume production.

WNERE? – Switching circuits in general. The 2N1252 and 2N1253 are ideally suited to high-speed high-current switching applications such as magnetic-core drivers, drum and tape write drivers, high-current pulse generators and clock amplifiers. In addition, the transistors are applicable to medium-speed saturated logic circuits.

Symbol	Characteristic	Rating	Min	Тур	Max	Test Co	nditions
hFE	D.C. pulse current gain 2N1252 2N1253		15 30	35 45	45 90	I _C =150mA	V _C =10V
PC	Total dissipation at 25°C case temperature	2 watts					
VBE SAT.	Base saturation voltage			0.9V	1.3V	Ic=150mA	lg = 15mA
VCE SAT.	Collector saturation voltage			0.6V	1.5V	I _C ==150mA	1 _B =15mA
h _{fe}	Small signal current gain at f=20mc 2N1252 2N1253		2 2.5	4 5.5		I _C =50mA	V _C =10V
СВО	Collector cutoff current			0.1µA 100µA	10μA 600μA	Vc = 20V	T 25°C
ts+tf	Turn off time			7 5mμs	150mµs	IC=150mA	I _{B1} -= 15mA
						I _{B2} =5mA	R _L =40Ω
						Pulse width=	10ms

FAIRCHILD 2N1252 and 2N1253

For full specifications, write Dept. E.1 Size as at Applitud2101-3-57 on the second floor at IRE



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(Continued from page 36.4)

associate directors of research.

According to J. B. Dow, president of this subsidiary of Hazeltine Corporation, Mr. Bailey will be responsible for military electronic apparatus research: Mr. Farber will direct the industrial research division and Mr. Richman will head the systems research division. These are new divisions established recently by Hazeltine Research Corporation.

Mr. Bailey holds M.E. and M.S. degrees (1933 and 1941) from Stevens Institute of Technology, where he taught before joining Hazeltine in

1936. Until his appointment, he had been chief engineer of the research division of Hazeltine Research Corporation. In 1954-55, Mr. Bailey was chairman of the Long Island section of the IRE. He is currently chairman of the IRE Com-



W. F. BAILEY

mittee on Television Systems and a member of the IRE Standards Committee. His articles and papers have appeared in various electronic publications.

An honor graduate of Columbia University in 1944, Mr. Farber received the M.E. degree from New York University in





R. J. FARBER

IRE's Radio Interference Committee and secretary of both the Broadcast Television Systems Committee and the National Stereophonic Radio Committee (Panel 1) of the Electronic Industries Association. He has authored many papers and articles in the fields of radio and television.

Winner of the IRE 1957 V. K. Zworykin Television Prize, Mr. Richman received the B.E.E. degree from the City College of New York in 1013

New York in 1943 and the M.E.E. degree from Brooklyn Polytechnic Institute in 1948. He joined Hazeltine in 1943 and has been chief engineer of the consultig andspecial studies division. He has held responsible positions in the company's IFF, FM,



D. RICHMAN

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(Continued on fage 40A)



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38A



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CONTINUOUS DEMONSTRATIONS AT BOOTH 4231 IRE SHOW



(Continued from page 38A)

super-regeneration, monochrome and color television, direction finding and other programs. He is the author of numerous articles and papers on communications theory, information theory, color systems, superregeneration, radiation theory and other electronics fields.

All three associate directors of research have been awarded U.S. and foreign patents for their work in electronics.

Hazeltine Research Corporation is planning to expand the staffs of its research divisions, which are now moving into the company's new research and development laboratory in Plainview, Long Island, Mr. Dow said. The establishment of divisions for systems research, military electronic apparatus and circuit research and commercial research are part of a program which will develop creative ideas in the field of communications, navigation, guidance and industrial research demanded by the space age and the expanding electronics needs of industry, Mr. Dow added.

Lawrence S. Churchill, Jr. (M'57) has joined Stavid Engineering, Inc. as engineering consultant in underwater electro-

magnetic propagation and ASW projects. He was formerly a member of the technical staff of Bell Telephone Laboratories, Inc., where he was engaged in research and development. and systems engineering in connection with under-



L.S. CHURCHILL, JR.

sonar systems. He received the B.S.M.E. degree from the Massachusetts Institute of Technology and the M.S.M.E from the University of Louisville.

He served on the faculty of the University of Louisville as an instructor in mechanics, mechanical design, vibration theory, fluid mechanics, and mechanical laboratory courses, and was Assistant Director of the university's Institute of Industrial Research. Mr Churchill is a member of the American Society of Mechanical Engineers.

Ivan S. Coggeshall (A'26-M'29-F'42) has been appointed to the Headquarters Staff of the American Institute of Electri-

cal Engineers as Manager Of Technical Operations Services. He served as Assistant Vice-President to Western Union Telegraph Company until his retirement on January 1, 1960.

Born September 30, 1896 in Newport, R I., he attended Worcester



I. S. Coggeshall

(Continued on page 42A)

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(Continued from page 40.4)

Polytechnic Institute, Worcester, Mass. from 1914 to 1917. He received an honorary Doctor of Engineering degree from that institution in 1951. He graduated from the U.S. Navy Steam School, Stevens Institute of Technology, Hoboken, N. L. in 1919. In 1937 he passed the Professional Engineering examination of New York State, and in 1955 he received a Certificate in International Law from the Naval War College in Newport, R. I.

Since 1917, Dr. Coggeshall has worked continuously for Western Union Telegraph Co., except for a year of military service during World War I. He was Director of the Mexican Telegraph Co, from 1939 to 1953, and served as General Traffic Manager, International Communications, from 1946 to 1951. He has served as Ensign. Lieutenant Commander and Commander in the U. S. Naval Reserve Force, and served on the Board of War Communications, Submarine Cables, in World War II.

He has served on several Governmental Boards and Committees; in 1959 he was appointed to the National Industrial Advisory Committee of the Federal Communications Committee. From 1947 to 1959 he was a member of the Western Union interdepartmental Committee on Technical Publication, and associate editor of the "Western Union Technical Review."

(Continued on base 46.1)



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... for measuring harmonics and noise from 30 cycles to 100 kc

FREQUENCY RANGE: 30 cycles to 100 kc., ± 300 cycle vernier.

VOLTAGE RANGE: 100 microvolts to 300 volts full scale.

- BANDWIDTH: -3 db at 10 and 30 cycles round top; 100 cycles and 1 kc. flat top.
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28 Fields of Special Interest-

The 28 Professional Groups are listed below, together with a brief definition of each, the name of

Aeronautical and Navigational Electronics Annual fee: \$2. The application of electronics to opera- tion and traffic control of aircraft and to navigation of all craft. Mr. Lewis M. Sherer, Chairman, RTCA, Washington, D.C. 33 Transactions, *5, *6, & *0, and *Vol. ANE-1, Nos. 2 and 3; Vol. 2, No. 1.3; Vol. 3, No. 2; Vol. 4, No. 1, 2, 3; Vol. 5, No. 2, 3, 4; Vol. 6, No. 1, 3.	Antennas and Propagation Annual fee: \$4. Technical advances in antennas and wave propagation theory and the utili- sation of techniques or products of this field. Mr. Arthur Dorne, Chairman, Dorne & Margolin, Westbury, L.L., N.Y. 27 Transactions, Vol. AP-2, No. 2; AP-4, No. 4; AP-5, No. 1-4; AP-6, No. 1, 2, 3, 4; AP-7, No. 1, 2, 3, 4.	Annual fee: 52. Technology of communication at audio frequencies and of the audio portion of radio frequency systems, including acoustic terminations, recording and reproduction. Dr. A. B. Bereskin, Chairman, EE Dept., Univ. of Cincinnati, Cincin- nati 21, Ohio. 49 Transactions, "Vol. AU-1, No. 61, "Vol. AU-2, No. 1, 4; Vol. AU-3, No. 1, 3, 5; Vol. AU-4, No. 1, 5-6; Vol. AU-5, No. 1, 2, 3, 4, 5, 6; AU-6, No. 1, 2, 3, 4, 5, 6; AU-7, No. 1, 2, 3, 4, 5.
Automatic Control Annual fee: \$2. The theory and application of auto- matic control techniques including feedback control systems. Mr. John E. Ward, Chairman, Servo- mechanisms Lab., MIT, Cambridge 39, Mass. 7 Transactions, PGAC-3-4-5-6, AC-4, No. 1.	Broadcast & Television Receivers Annual fee: \$2. The design and manufacture of broad- cast and television receivers and com- ponents and activities related thereto. Mr. Robert R. Thalner, Chairman, Sylvania Home Electronics, Batavia, N.Y. 23 Transactions, *7, 8; BTR-1, No. 1-4; BTR-2, No. 1-2-3; BTR-3, No. 1-2; BTR-4, No. 2, 3-4; BTR-5, No. 1, 2.	Broadcasting Annual fee: \$2. Broadcast transmission systems engi- neering, including the design and utili- zation of broadcast equipment. Mr. George E. Hagerty, chairman, Westinghouse, 122 E. 42nd St., New York 17, N.Y. 14 Transactions, No. 2, 4, 6, 7, 8, 9, 10, 11, 12, 13, 14.
Circuit Theory Annual fee: \$3. Design and theory of operation of cir- cuits for use in radio and electronic equipment. Mr. Sidney Darlington, Chairman, Bell Tel. Labs., Murray Hill, N.J. 25 Transactions, CT-2, No. 4; CT-3, No. 2; CT-4, No. 3-4; CT-5, No. 1, 2, 3, 4, CT-6, No. 1, 2, 3.	Communications Systems Annual tee: \$2. Radio and wire telephone, telegraph and facsimile in marine, aeronautical, radio-relay, coaxial cable and fixed sta- tion services. Mr. J. E. Schlaijker, Chairman, IT&T, 67 Broad St., New York 4, N.Y. 15 Transactions, CS-2, No. 1; CS-5, No. 2, 3; CS-6, No. 1, 2; CS-7, No. 1, 2, 3.	Component Parts Annual fee: \$3. The characteristics, limitation, applica- tions, development, performance and re- liability of component parts. Mr. J. J. Drvostep, Chairman, Sperry Gyroscope Co., Great Neck, N.Y. 18 Transactions, Vol. CP.3, No. 2; CP-4, No. 1, 2, 3-4; CP-5, No. 1, 2, 3, 4; CP-0, No. 1, 2, 3, 4.
Education Annual fee: \$3. To foster improved relations between the electronic and affiliated industries and schools, colleges, and universities. Dr. John G. Truxal, Chairman, Dept. of EE, PIB, Brooklyn, N.Y. 8 Transactions. Vol. E-1, No. 3, 4; E-2, No. 1, 2, 3, 4.	Electron Devices Annual fee: \$3. Electron devices, including particularly electron tubes and solid state devices. Dr. W. M. Webster, Chairman, RCA Labs., Princeton, N.J. 26 Transactions, *Vol. ED-1, No. 3-4; ED-3, No. 2-4; ED-4, No. 2-3, 4; ED-5, No. 2, 3, 4; ED-6, No. 1, 3.	Electronic Computers Annual fee: \$4. Design and operation of electronic com- puters. Mr. Richard O. Endres, Rese En- gineering Co., 731 Arch St., Phila- delphia, Pa. 31 Transactions, EC-6, No. 2, 3; EC-7, No. 1, 2, 3, 4; EC-8, No. 1, 2, 3.
Engineering Management Annual fee: \$3. Engineering management and adminis- tration as applied to technical, indus- trial and educational activities in the field of electronics. Dr. Henry M. O'Bryan, Sylvania Elec. Products, 730 3rd Ave., New York 17, N.Y. 16 Transactions, EM-3, No. 2, 3; EM-4, No. 1, 3, 4; EM-5, No. 1-4; EM-6, No. 1, 2, 3.	Engineering Writing and Speech Annual fee: \$2. The promotion, study, development, and improvement of the techniques of preparation, organization, processing, editing, and delivery of any form of information in the electronic-engineer- ing and related fields by and to in- dividuals and groups by means of direct or derived methods of communication. Mr. T. T. Patterson, Jr., Chairman, RCA Bldg. 13-2, Camden, N.J. 4 Transactions, Vol. EWS-1, No. 2; EWS-2, No. 1, 2.	Human Factors in Electronics Annual fee: \$2. Development and application of human factors and knowledge germane to the design of electronic equipment. Mr. Curtis M. Jansky, Chairman, Royal McBee Corp., Port Chester, N.Y.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

4.

-IRE's 28 Professional Groups

the group chairman, and publications to date.

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* Indicates publications still available

Industrial Electronics Annual fee: \$3. Electronics pertaining to control, treat- ment and measurement. specifically, in industrial processes. Mr. J. E. Eiselein, Chairman, RCA Victor Dev., Camden, N.J. 10 Transactions, *PGIE-1-3-5-6-7-8, 9, 10.	Information Theory Annual fee: \$3. Information theory and its application in radio circuitry and systems. Dr. Peter Elias, Chairman, MIT, Cambridge 39, Mass. 18 Transactions, PGIT-4, IT-4, No. 2-3; IT-2, No. 3; IT-3, No. 1, 2, 3, 4; IT-4, No. 1, 2, 3, 4; IT-5, No. 1, 2, 3	Instrumentation Annual fee: \$2. Measurements and instrumentation uti- lizing electronic techniques. Mr. C. W. Little, Jr., Chairman, C-Stellerator Assoc., Princeton, N.J. 16 Transactions, 4: Vol. 1-6, No. 2, 3, 4: Vol. 1-7, No. 1, 2: Vol. 1-8, No. 1, 2, 3.
Medical Electronics Annual fee: \$3. The use of electronic theory and tech- niques in problems of medicine and biology. Mr. W. E. Tolles, Chairman, Air- borne Instruments Lab., Mineola, N.Y. 15 Transactions, 8, 9, 11, 12, ME-6, No. 1, 2, 3.	Microwave Theory and Techniques Annual fee: \$3. Microwave theory, microwave circuitry and techniques, microwave measure- ments and the generation and amplifica- tion of microwaves. Dr. A. A. Oliner, Microwave Re- search Institute, 55 Johnson St., Brooklyn 1, N.Y. 27 Transactions, MTT-4, No. 3-4; MTT-5, No. 3, 4; MTT-6, No. 1, 2, 3, 4; MTT-7, No. 2, 3, 4.	Military Electronics Annual fee: \$2. The electronics sciences, systems, ac- tivities and services germane to the re- quirements of the military. Aids other Professional Groups in liaison with the military. Mr. Henry Randall, Chairman, Office of Asst. Secy. Defense, Pentagon, Washington, D.C. 7 Transactions, MHL-1, No. 1; MHL-2, No. 1; MHL-3, No. 2, 3, 4
Nuclear Science Annual fee: \$3. Application of electronic techniques and devices to the nuclear field. Dr. A. B. Van Rennes, Chairman, United Research, Inc., Cambridge, Mass. 14 Transactions, NS-1, No. 1; NS-3, No. 2, 3; NS-4, No. 2; NS-5, No. 1, 2, 3, NS-6, No. 1, 2, 3.	Production Techniques Annual fee: \$2. New advances and materials applica- tions for the improvement of produc- tion techniques, including automation techniques. Mr. L. M. Ewing, Chairman, Gen- eral Electric Co., Syracuse, N.Y. 6 Transactions, No. 2-3, 4, 5, 6.	Radio Frequency Techniques Annual fee: \$2. Origin, effect, control and measurement of radio frequency interference. Mr. J. P. McNaul, Chairman, Signal Corps, USA's RDL, Ft. Monmouth, N.J. 1 Transaction, RF-1, No. 1.
Reliability and Quality Control Annual fee: \$3. Techniques of determining and con- trolling the quality of electronic parts and equipment during their manufac- ture. Mr. P. K. McElroy, Chairman Gen- eral Radio Co., West Concord, Mass. 16 Transactions, *3, 5-6, 10, 11, 12, 13, 14, 15, 16.	Space Electronics and Telemetry Annual fee: \$2. The control of devices and the meas- urement and recording of data from a remote point by radio. Mr. C. H. Hoeppner, Chairman, Ra- diation, Inc., Melbourne, Fla. 14 Transactions, TRC-1, No. 2-3; TRC-2, No. 1; TRC-3, No. 2, 3; TRC-4, No. 1; SET-5, No. 1, 2, 3, 4.	Ultrasonics Engineering Annual fee: \$2. Ultrasonic measurements and communi- cations, including underwater sound, ultrasonic delay lines, and various chemical and industrial ultrasonic de- vices. Dr. Wilfred Roth, Chairman, Roth Lab., Hartford, Conn. 7 Transactions, PGUE, 5, 6, 7.
Vehicular Communications Annual fee: \$2. Communications problems in the field of land and mobile radio services, such as public safety, public utilities, rail- roads, commercial and transportation, otc. Mr. A. A. MacDonald, Chairman, Motorola, Inc., 4545 W. Augusta Blvd., Chicago 51, 111.	USE THIS Miss Emily Sirjane IRE—1 East 79th St., New York 21, N.Y Please enroll me for these IRE Professio Name Address Place Please enclose remittance with this ord	COUPON PG-3-60 7. onal Groups \$ \$ \$ er.



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IRE People

(Continued from page 42.41)

He has also authored several engineering papers and articles, and written many book reviews.

Dr. Coggeshall is a Fellow of the AIEE and a member of Tau Beta Pi. He is a Past President and Director of the IRE.

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Robert B. Corby (M'56) has been appointed to the position of Staff Engineer in the Program Planning Department of Motorola's Western

Military Electronics Center, it was announced by Harvey M. Ross, Program Planning Manager.

In his new position he will be responsible for the development of new areas of product and program activities. This will include the analysis



R. B. Corby

include the analysis, organization and coordination of technical and promotional effort required to match Motorola's skills most appropriately to the military's needs.

He joined Motorola in 1953 as Assistant Manager in Microwave Products in Chicago, Ill. In 1955 he became Manager of Military Microwave Sales and in 1958 was transferred to the Military Electronies Division in Phoenix, Ariz, as marketing coordinator for the company's six military plants. Prior to joining Motorola, Mr. Corby had extensive electronics engineering and administrative experience and during World War II was Navy liaison representative to the M.I.T. Radiation Laboratory. A graduate of Union College with the B.S.E.E. degree in 1941, he holds the M.S. degree in marketing from the University of Chicago.

Dr. Lloyd T. DeVore (A'42-SM'44-F'52) has been appointed director of engineering of the Laboratories Division, Hoffman Electronics Corp., President II. Leslie Hoffman announced.

The appointment was made to coordinate more effectively the division's research and engineering activity with advanced research at the Hoffman Science Center in Santa Barbara, Calif., which Dr. DeVore also heads.

The new engineering director joined Hoffman a year ago, as a corporate vice president and director of the Science Center, which engages in advanced, product oriented research in some of the promising new fields in electronics.

He began his career as a physics instructor at Pennsylvania State College after receiving the B.S. degree in 1930, the M.S. degree in 1931 and the Ph.D. degree in 1933. In 1942 he became chief engineer

(Continued on page 48.4)

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Multicoated for extra long wear by a new exclusive process. Solder adheres only to working surface at point of tip — prevents solder dropping on components or creeping into tip hole. Eliminates costly tip maintenance.

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



HEWLETT PACKARD Specifies Tung-Sol tubes for high stability calibration generator

The Hewlett-Packard Voltmeter Calibration Generator calibrates high impedance voltmeters and oscilloscopes with extreme accuracy. An exceptionally stable source for a wide range of precision voltages, the premium instrument speeds up production and maintenance testing.

To assure high stability and low distortion performance, which are listed among the unit's principal advantages Hewlett-Packard selected Tung-Sol 6550's for the 400 cycle power amplifier. As Hewlett-Packard reports: "Tung-Sol's 6550 shows unusual insensitivity to load changes."

What this means, of course, is that under varying loads the 6550 drive, with its tight characteristics, holds to a minimum any change in the unit's already minimal distortion (less than 0.2%). In addition the 6550 helps to provide long-term stability.

Like all Tung-Sol components, the 6550's optimum performance and dependability stems from Tung-Sol's deep-rooted component know-how. Every step in the manufacturing process is carefully disciplined. Stringent quality control guarantees uniformly high performance in any one lot or from lot to lot. And exhaustive life tests under severe overload assures adequate safety margins.

Maybe you're up against some exacting component requirements. If so, you'll be steering a wise course by getting in touch with Tung-Sol applications engineers. They're component experts who will gladly study your design and recommend the units that will do the job . . . precisely. Tung-Sol Electric Inc., Newark 4, New Jersey. TWX:NK193.

For prompt and competent technical consultation on Tung-Sol components call the Tung-Sol Commercial Engineering office near you. SALES OFFICES: Atlanta, Ga.; Columbus, Ohio; Culver City, Calif.; Dallas, Texas; Denver, Colo.; Detroit, Mich.; Irvington, N. J.; Melrose Park, Ill.; Newark, N. J.: Philadelphia, Pa.; Seattle, Wash. Canada: Montreal, P. Q.





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(Continued from page 46.4)

of the Special Projects Laboratory, Wright Field, specializing in radio and radar communications and in guided missiles. After World War II he joined the University of Illinois, where he became chairman of the Research Committee and coordinator of research for the Department of Electrical Engineering.

Dr. DeVore left the university in 1950. to become manager of General Electric Company's Electronics Laboratory in Syracuse, N. Y., serving five years before accepting an assignment as general manager of the Electronics Division, Stewart-Warner Corporation.

He is a fellow of the American Association for the Advancement of Science, and a member of the American Physical Society, American Management Association and of the Armed Forces Communications and Electronics Association.

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Airborne -Instruments Laboratory (AIL), Deer Park, N. Y., yesterday announced the appointments of Dr. E. G. Fubini (A'36-SM'46-F'54) and Dr. G. C. Comstock (A'47-M'47-SM'51-F'57) as vice presidents. Dr. Fubini and Dr. Comstock formerly were co-directors of AIL's Research and Engineering Division, AIL is a division of Cutler-Hammer, Inc.

Dr. Comstock was appointed vice president of the Electronic Systems and Techniques Division of Airborne Instruments Lab.

He has been concerned with the system development of radars and radar data utilization since early when 1942. he joined the staff of the Radiation Laboratory, Massachusetts Institute of Technology. Λt that time, his effort



G. C. COMSTOCK

was devoted to the early development of instrument landing radar systems.

He received the B.S. degree at Bradley University in 1932 and the Ph.D. degree in Physics from the University of Chicago in 1938, and served as Assistant Professor of Physics at the Citadel until 1942, when he went to M.L.T.

At M.I.T. in 1943, he was in charge of the electronics engineering for the original GCA (Ground Controlled Approach) radar. Subsequently, he served as GCA Project Leader and an Associate Chairman of the Ground Radar Division during the initial development of airport traffic control radar.

In 1946, shortly after joining AIL, he became Assistant Supervising Engineer of the Air Navigation and Traffic Control Section, and in 1950 Supervising Engineer of the Operational Development Section and later of the Radar and Navigation Section. From 1955 through 1959, he was

(Continued on page 54A)

WHEN WRITING TO ADVERTISERS PLEASE MENTION ---- PROCEEDINGS OF THE IRE

Semiconductor News from SYLVANIA Quality-by intention!

Sylvania NPN and PNP Transistors controlled specifically for switching service

Rigid adherence to high standards of performance and electrical uniformity is assured through the exercise of stringent quality controls. High reliability under severe environmental conditions is assured by thorough final-test procedures. Sylvania switching transistors are in TO-5 cases with welded hermetic seal. Shown here are a number of switching circuits designed around Sylvania transistors and diodes.

"NAND" diode-transistor gates

"NOR"

diode-transistor gates



SYLVANIA NPN AND PNP SWITCHING TRANSISTORS

Reliable performers in military and computer applications

1	ELECTRICAL CHARACTERISTICS					
NPN Type	COLLECTOR TO BASE VOLTS (Min.)	EMITTER TO BASE VOLTS (Min.)	POWER DISS. AT 25°C (Max.)	FREQ. CUTOFF, FAB V _{cB} =6v. 1c=1ma (Min.)		
	151	151/	100mW	3.0Mc		
2N312	150	201	100mW	3.0Mc		
2N356	200	200	100mW	6.0Mc		
2N357	20V	201	100mW			
2N358	20V	200	150mW	2.5Mc		
2N377	25V	150	200mW	2.5Mc		
2N377A	40V	150	150mW	4.0Mc		
2N385	25V	15V	200mW	4.0Mc		
2N385A	40V	15V	150mW	5.0Mc		
2N388	25V	15V	150mW	5.0Mc		
2N3884	40V	15V	200mw	2.5Mc		
2N/38	30V	25V	100mW	2.5Mc		
211438	30V	25V	150mW	5 OMc		
211438A	30V	25V	100mW	5.0Mc		
2N439	307	25V	150mW	5.000		
2N439A	307	25V	100mW	10.0Mc		
2N440	300	251	150mW	10.0Mc		
2N440A	300	101	100mW	- 1		
2N556	25V	107	100mW			
2N557	20V	100	100mW	-		
2N558	15V	57	200mW	5.0Mc		
2N576	20V	15V	200mW	5.0Mc		
2N576A	40V	150	120mW	3.0Mc		
2N585	25V	20V	150-14	_		
2N587	40V	40V	150mw	2 0Mc		
2N679	25V	15V	150mW	2.0Mc		
2N1202	25V	25V	150mW	5.0Mc		
2N1302	25V	25V	150mW	5.0MC		
211304	25V	25V	150mW	TU.UMC		
2N1306	251	25V	150mW	15.0Mc		
2N1308	257	15V	150mW	7.0Mc		
2N1114	250	15V	150mW	4.0Mc		
2N1299	400	151				
PNP Type	COLLECTOR TO BASE VOLTS (Min.)	EMITTER TO BASE VOLTS (Max.)	POWER DISS. AT 25°C (Max.)	FREQ. CUTOFF, FAB V _{c9} =5 1e=1mA (Min.)		
		101/	150mW	5.0Mc		
2N123	_20V	-104	150mW	4.0Mc		
2N404	_25V	$1 - \frac{12}{12}$	150mW	5.0Mc		
2N414	_30V	-12V	150mW	2.5Mc		
201425	30V	_20V	150mW	3.0Mc		
211425	_30V	_20V	150mW	5.0Mc		
211420	_30V	_20V	150000	10.0Mc		
	_30V	20V	150mw	0.5Mc		
2N428	_25V	_15V	150mW	14 OMC		
2N519	_25V	-12V	150mW	0.5Mc		
2N582	_10V	-	120mW	0.5Mc		
2N1009	25V	_15V	150mW	U.SIVIC		
2N1381	-254			1		

SYLVANIA 2N624 "DRIFT" TRANSISTOR FOR TUNED-AMPLIFIER SERVICE TO 12.5 MC

Sylvania 2N624 is a hermetically sealed PNP diffused-base transistor. The package has JEDEC TO-12 dimensions and lead spacings. A fourth lead provides a connection to the metal case for improved shielding. Characteristic testing includes many environmental parameters to assure reliable operation under conditions which may be expected in military applications. Sylvania 2N624 conforms to the requirements for military electronics equipment.



SYLVANIA DIODES—Sylvania manufactures all types of diodes for service as gates, clippers, clampers, detectors; diodes for applications in communications equipment, switching circuits in electronic computers operating at high speeds in the order of millimicroseconds, and special-purpose electronic devices.

> SYLVANIA facilities for life and environmental testing include salt spray, moisture, high altitude, Vibration, shock, high and low temperatures. SYLVANIA manufacturing and testing facilities are highly automated and mechanized to assure extraordinary electrical uniformity. Many SYLVANIA diodes are available with specifications conforming to military requirements.

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feature low cost, low capacitance, and exceptionally fast recovery time. Available in all-glass "min" package with power dissipation capabilities to 80mW. Available in solder-seal package for wire-in or clip-in use with power dissipation capabilities to 225mW.



GOLD BOND DIODES

feature high forward-conduction and good recoverytime in units that are relatively low in cost. Available in all-glass "min" package with power dissipation capabilities averaging 80mW.

VLI (very low impedance) DIODES

feature very high conduction and relatively high voltage-breakdown. Available in all-glass "min" package with power dissipation capabilities averaging 80mW. Available in solderseal package for wire-in or clip-in use with increased power dissipation capabili-

SILICON-JUNCTION DIODES



feature high conduction, good recovery time plus the environmental capabilities of silicon-the ability to withstand wide variations in ambient temperature. Available in all-glass "min" package with power dissipation capabilities to 200mW.

SYLVANIA D-1820 HIGH-SPEED SWITCHING DIODE 4 millimicroseconds guaranteed maximum recovery time!

ties to 225mW.

Absolute Maximum Ratings*					
Fwd. V	'olt	1.3 V †			
Fwd. C	Curr	50 mA			
Back V	'olt	20 V			
Pwr. D	'iss	80 mW			

Typical Operating Conditions®
Fwd. Volt

tat 10 mA at 20° C.

SYLVANIA D-1820—now available in commercial quantities—is designed, produced and controlled specifically for logic circuitry. The cost of this SYLVANIA diode is low enough to make it especially attractive for use in quantityproduced electronic computers. SYLVANIA D-1820 and circuits designed around it feature: high-speed operation • long-life performance • high reliability • exceptional uniformity • economy • simplicity • compactness.

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SYLVANIA silicon rectifiers are quality-controlled for applications in *industrial power supplies* and *magnetic amplifiers*. SYLVANIA silicon rectifiers are available with peak-inverse-voltage ratings to 1000-Volts, and forward-current ratings to 750-mA.



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SYLVANIA semiconductor devices are available from your local franchised SYLVANIA SEMICONDUCTOR DISTRIBUTOR or through the FIELD OFFICE nearest you. For technical data, write: SYLVANIA SEMICONDUCTOR DIVISION, WOBURN, MASSACHUSETTS.



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IRE People

(Continued from page 48A)

co-director of the Research and Engineering Division of AIL

Since 1949, he has also served on many Department of Defense research and development committees and working groups. He has been a member of the Research and Development Board's Panel on Air Navigation and chairman of the subpanel on Short-Distance Navigation Aids, chairman of an RDB Working Group on Counter-Countermeasures for the Guidance and Control of Guided Missiles, a member of the Air Defense System Engineering Committee of the Air Force Scientific Advisory Board, participant in the Project Charles Air Defense Study, consultant for the Technical Advisory Panel on Electronics to the Department of Defense, served with the Weapons Systems Evaluation Group, Office of the Secretary of Defense on special defense studies, and most recently on the Air Defense Panel of Presidential Scientific Advisory Committee.

Dr. Comstock has been active with the Navigation Aids Committee of IRE and is presently serving as secretary-treasurer of the Professional Group on Engineering Management.

Dr. Fubini was appointed vice president of the Research and Systems Engineering Division of Airborne Instruments

Lab. During the period 1942 to 1945. as a Research Associate of the Harvard University Radio Research Lab., he was concerned with the design, development. operation. and planning of countermeasure and ferret reconnaissance equipment. During



E. G. FUBINI

1943 and 1944 he was a Scientific Consultant and Technical Observer to the U.S. Army and U. S. Navy in the European Theatre of Operations, where he participated in the establishment of electronic reconnaissance and jamming capabilities for the invasion of Italy and of Southern France.

During 1944 and 1945 he was in England with the U.S. Eighth Air Force in charge of electronic reconnaissance and countermeasures. During 1945 he was a special consultant for ECM to the Office of the Air Communication Officer of the War Department.

He joined AlL as an engineer in 1945, where he was concerned with the development of microwave components, magnetic detectors, electronic test equipment, boundary value problems, AJ devices, antennas, direction finders, and reconnaissance systems.

Under his supervision, AIL became involved in such developments as APR-9 Receivers, ARA-25 Direction Finders, ASQ-8 Magnetic Airborne Detectors, Around-the-Mast Rotary Joints, AN

(Continued on Page 56.4)

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



The burning question of cool flames

Between the brief stage of not burning and burning, many hydrocarbons react with oxygen at temperatures well below that of normal flame combustion. But the reactions are usually transient and hard to analyze. At the General Motors Research Laboratories, we have been able to investigate the effect of chemical additives on cool preflames.

To do this, the almost invisible cool flames are stabilized for hours in a flat-flame burner, permitting careful examination of the retardation or acceleration effects of the additives. From more than twenty additives studied, experimental results indicate that some chemicals affect combustion through the mechanism of preflame reactions. We are now accumulating new information on these additives' mode of operation. For instance: emission spectra support the conclusion that tetraethyl lead reacts with the oxygenated compounds formed in cool flames to vield lead oxide vapor. These findings of when and how lead oxide is formed are important in resolving a current controversy of science the combustion behavior of tetraethyl lead.

Studies such as this may lead to more economical and effective means of controlling unrestrained combustion - such as "knock" in reciprocating engines. The work is typical of GM Research's effort to provide useful information for a moving America. And in this way continue to keep our promise of "More and better things for more people."

General Motors Research Laboratories Warren, Michigan



Iron carbonyl, an antiknock



Ethyl nitrate, a proknock

Iron carbonyl retards. ethyl nitrate accelerates central portion of cool flames.



Sensitive aircraft and missile components and systems often require temperature control within close limits – while ambient temperatures fluctuate widely. Eastern refrigeration-type cooling systems are ideal for such conditions.

Designed for the strictest military requirements, these vapor-cycle closed-system packages are built around a highly efficient compressor powered by a special 400-cycle motor. Unique condensing and special cooling methods are called upon to meet the most unusual operating requirements, the most demanding specifications.

Capacities range from 100 to 6000 watts; operating altitudes extend to 100,000 feet. Some units, of the "boil-off" type, perform almost without regard for extremes in altitude and temperature.

Call on Eastern for imaginative solutions to *all* avionic cooling problems . . . and write for new Bulletin 360.





other refrigeration units for aircraft and missile electronics



BUILD ON . .



IRE People

(Continued from page 54.4)

/ASD-1 System and 117L Subsystem.

Dr. Fubini obtained his Doctor of Physics degree in Rome, Italy, A former Lecturer of Harvard University, he has received the Presidential Certificate of Merit. He is the author of about thirty technical publications and holds eleven patents. He is a member of the Air Force Scientific Advisory Board, chairman of the Advisory Group on Electronic Warfare of the Department of Defense, member of the Advisory Group on Special Projects of the Department of Defense, chairman of the Electromagnetic Warfare Advisory Group of the Air Research and Development Command, member of a panel of the Scientific Advisory Committee to President Eisenhower, member of a panel of the National Security Agency Scientific Advisory Board, and member of the Advisory Council for the Advancement of Scientific Research and development in New York State.

Floyd M. Gardner (S'49–A'54-SM'58) has announced the formation of Gardner Research Company. The new firm will provide electronics consulting services to government and industry and will be located at 9881 Nichols, Orange, Calif. Dr. Gardner was formerly associate director of Research at Interstate Electronics Corporation.

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Appointment of **Robert J. Gilson** (SM'55) as director of systems management in Stromberg-Carlson's Electronics Division has been

announced by Kenneth M. Lord, vice president and general manager of the division. Stromberg-Carlson is a division of General Dynamics Corp.

Mr. Gilson comes to Stromberg-Carlson from the Hoffman Laboratories Division of Hoff-



R. J. GILSON

man Electronics Corp., Los Angeles, Calif., where he was program director for the "Tall Tom" airborne reconnaissance system. Prior to his association with the Hoffman company Mr. Gilson was with Litton Industries, in Beverly Hills, Calif., for over two years, and earlier, with the Sylvania Electronic Defense Laboratory in Mountain View, Calif.

A native of Palo Alto, Calif., Mr. Gilson received the B.S. degree in electrical engineering from Montana State College, and the M.S. degree in engineering administration from the University of California at Los Angeles. During World War II he served as a radio officer in the Maritime Service. He is a member of the American Institute of Electrical Engineers, the

(Continued on page 58A)



WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Umbilical

The MSC-built Umbilical Launching Cable...an example of the product diversity of Missile Systems Corporation. Like all products that bear the MSC label, this system has proven its reliability. Just as it is a life-line to the success of a mission, so also are MSC's contributions material to the future accomplishments of *all* facets of the electronic industry. MSC's variety of products form one continual life-line...feeding an industry which is already changing the life patterns of generations to come.



MISSILE SYSTEMS CORP. 11949 VOSE ST., NORTH HOLLYWOOD, CALIF.

Engineering and Manufacturing Corporation of Texas/subsidiary Dallas, Texas



A new series of completely transistorized I-F amplifiers offered to fill the need for standardized, high quality units. These T-330 series amplifiers by 1.F.I. are avail-able in a variety of center frequencies and bandwidths. They also can be equipped with emitter follower, cathode detector or low noise tube input.

The quality of construction is high. The use of printed circuitry and quality control procedures provide rigid standards. Indi-vidual inspection and testing of each unit prior to delivery assure the superior quality of IFI transistorized I-F amplifiers. These transistorized ampli military environment

	Unit
antity	Price
10	\$800
-25	700

ifiers meet all applicable ntal specifications,	Input Impedance
Unit Price	Noise Figure
\$800 700	Mean Stage Gain

11 T-330A T-330B **Mean Stage Gain** INSTRUMENTS FOR INDUSTRY, Inc. 101 New South Road, Hicksville, L. I., N.Y.

Graduate engineers with two or more years of circuit application in the fields of electronics or physics are invited to most with Mr. John Hicks in an informal interview or send complete resume to: Dir. Personnel, IFI, 101 New South Road, Hicksville, New York. John Hicks in an informal interview v or send complete resume to 1 Dir. Personnel, IFI, 101 New 3 See us at the IRE Show—Booth 1424



(Continued from page 56.4)

American Management Association, the American Institute of Management and the Association of the U.S. Army.

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Dr. John L. Grigsby (SM'56) has joined the Palo Alto advanced electronic systems firm of Applied Technology, Inc., as Chief Engineer.

He will direct the company's expanding engineering activities in reconnaissance receiving systems, active electronic countermeasures and special instrumentation for ionosphere and radio astronomy studies.

Formerly Dr. Grigsby was with the Stanford University Applied Electronics Laboratory, since 1952 serving successively as engineer, project engineer and group leader. He also served as a consultant to several electronic firms and was a memberof a special advisory committee to the U. S. Army.

In his new association he will continue his work in the application of travelingwave and related beam tubes in broadband countermeasures and receiving systems. This has been his principal work with the Stanford laboratory.

Previously with the General Electric Company from 1949 to 1952, he instructed military personnel and GE field service engineers in theory and operation of firecontrol, long-range search and heightfinding radars.

During World War II he served for two and a half years with the U.S. Air Force, mostly in electronics and radar work.

Dr. Grigsby received the B.S. degree in electrical engineering from the University of Colorado in 1948. Stanford University awarded him the M.S. and Ph.D. degrees in the same field in 1955 and 1959, respectively.

He is a member of Tau Beta Pi, Eta Kappa Nu, Sigma Xi and Sigma Tau.

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The appointment of John E. Hogg (S'39-A'42-M'46) as manager of marketing administration and personnel development for the Gen-

eral Electric Company's Computer Department has been announced by G. A. Hagerty, Department manager of marketing.

Mr. Hogg was formerly western regional manager for the Department at Palo Alto, Calif. He will transfer to



JOHN E. HOGG

Department headquarters in Phoenix, Ariz.

In his new position, he will be responsible for all phases of product availability and delivery, including contracts administration and marketing office management,

(Continued on page 62A)



Qu

SPECIFICATIONS

T-330/

330E

330E

330/

-330B

3304

330E

3304

1105

-330B

T-3304

Center Free

Bandwidth

Output (max)

Gain

30 mc 30 mc 10 mc

80 db min. 100 db min. + 5 DBM +10 DBM 50 ohm

50 ohm 10 db

9 db

14.0 D1



SILICON CONTROLLED RECTIFIERS

IN MINIATURE

actual size

FROM SOLID STATE

For control circuit application in the 10 to 1250 ma output current range

HIGH SENSITIVITY

only 2 mA input to control one ampere (continuous) at 100°C.

NOW.

- HIGH TEMPERATURE stable operation to 150°C.
- LOW LEAKAGE
 10 uA cutoff current at full voltage.
- SIMPLIFIED MOUNTING

no need for insulating hardware stud is electrically isolated.

Туре	Maximum Anode Voltage (DC or Peak AC)	Maximum Average Forward Current	Maximum Gate Current	Gate V to F + Vo	oltage ire olts
	± Volts	Amps	mA	Min,	Max.
38305	30	1.0	2	.40	2.5
3B60S	60	1.0	2	.40	2.5
3B100S	100	1.0	2	.40	2.5
3B150S	150	1.0	2	.40	2.5
3B200S	200	1.0	2	.40	2.5

These devices offer significant circuit advantages in that they are specifically designed for operation in the 10 to 1250 mA current range. It is no longer necessary to derate higher power units, with attendant losses in efficiency.

The miniature SCR combines a current rating of 1 ampere at 100°C with extremely small size. It features high peak recurrent and surge current ratings. Switching efficiency up to 98% is practical. High gain, low loss control of loads up to 300 watts can now be achieved along with significant miniaturization. The internally insulated junction eliminates the need for external mica washers. Assembly is therefore simplified and reliability improved.

The miniature SCR is useful in applications such as AC and DC static switching, proportioning control, D.C. to D.C. converters, servo motor driving, squib firing, protective circuits, and related applications.

Encapsulated in the unique SSPI cold welded copper case, the SCR offers a high degree of mechanical ruggedness and long term reliability.

WRITE FOR BULLETIN C415-01

On display at IRE



Booth 1519 B



Missile Battery

Development of the Silvercel battery 3381R-2 for missile applications has been announced by **Yardney Electric Corp.**, 40–50 Leonard St., New York, N. Y., manufacturer of silver-zinc and silvercadmium batteries.



A rechargeable silver-zinc power pack, this 10-ampere-hour unit has a nominal voltage of 28 volts when discharging at 45 amperes in 12 minutes. It can also be discharged at 60 amperes, or at lower rates. It has a volume of 239 cubic inches and weighs 15 pounds. Its dry shelf life is a minimum of two years.

Designed to meet requirements for missile electric power systems, the new battery has met test specifications of MIL E5272: up to 5 G's vibration; 15 G's, 11 milliseconds in all directions mechanical shock; -65° F low temperature; 160°F high temperature; 95% humidity at 160°F; 55,000 feet at 80°F high altitude.

16 Channel Recorder

Originally designed to verify magnetic tapes in machine tool mumerical control systems, the new multichannel recorders, a product of **Epsco**, a **Division of Epsco Incorporated**, 207 Main St., Worcester 8, Mass., also find industrial uses. These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

Through minor modification of existing equipment, the firm has been able to satisfy any graphic recording problems requiring more than eight channels on a common time basis.

Modular construction offers the maximum in versatility. Interchangeable preamplifiers and drivers permit rapid and economical change-over from one test to another.

The firm presently stocks five modular pre-amplifiers, low gain dc, moderate gain dc, high gain dc, ac, and carrier. Epsco's three stage differential input-output driver amplifiers are designed to supply driving power, frequency compensation and overload protection in damping impedance to the galvanometers. The ink or electric-writing galvanometers write on either standard roll chart or Epsco "Z" fold paper, and are housed in recorder cabinets equipped with nine selectable chart speeds.

Atmospheric-Particle Monitor

A new atmospheric-particle counting instrument has been designed for application by Royco Instruments, Inc., 365 San Antonio Rd., Mountain View, Calif., to the continuous monitoring of outdoor air or atmospheres of indoor locations, such as ultra-clean work areas. It presents, on a stripchart recorder, a permanent record of aerosols present in an overall range from 0.3 microns to any desired upper limit. This record is differentiated into 15 subranges and recorded in sequence at intervals which can be predetermined in length. Stability problems are eliminated by the fact that the unit is continuously self-calibrated.



Included in the instrument is an alarm system which can be set for a remote indication of particle concentrations in any of

the monitored sub-ranges exceeding a predetermined maximum.

Counting rate of the PC-200 is 1000 particles per minute with a 1 per cent coincidence loss at the standard flow rate of 100 cc per minute. Recordings are made on a 6-inch strip chart at a standard movement of 8 inches per hour. Operation is from a 115v 60cps supply with a current of 3 amperes.

Weighing approximately 150 lb, the unit is 21 by 19 by 32 inches overall. It is mounted on greaseless, dustless casters. Price \$6,975 f.o.b. Mountain View.

Precision Phase Detector

A new precision phase measuring instrument from 15 mc to 400 mc has been developed to meet the need of measuring phase shift of radar IF amplifiers, transmission networks and radar tracking systems by Ad-Yu Electronics Lab., Inc., 249 Terhune Ave., Passaic, N. J. This instrument utilizes a comparison method to achieve the accuracy of $\pm 0.05^{\circ}$ or $\pm 1\%$. The sensitivity has been increased with the use of balanced tuned amplifiers. In addition, balanced tuned amplifiers can also minimize the error due to harmonic contents and noise.



The accuracy is $\pm 0.05^\circ$ or $\pm 1\%$ of the dial reading. The resolution time is less than 0.1 micromicrosecond; the smallest phase angle which can be read on the dial is less than 10⁻¹³×360×frequency in cps. The time delay of the continuously variable delay line can be adjusted from 0 to 2.8 millimicroseconds. Two step variable delay lines have total delay of 37.5 millimicroseconds in E_1 channel and 7.5 millimicroseconds in E_2 channel (in steps of 1 millimicrosecond). The minimum input signal depends on the sensitivity of the receiver; approximately 20 microvolts for receiver having 5-microvolt sensitivity, and approximately 2 volts minimum is recommended for using panel meter as indicator. The characteristic impedance is 50 ohms nominal for both input and output; type N connectors are used throughout.



Creative Microwave Technology

Published by MICROWAVE AND POWER TUBE DIVISION, RAYTHEON COMPANY, WALTHAM 54, MASS., Vol. 1, No. 8

NEW RAYTHEON HEATERLESS AMPLITRONS EXCEED 1,000 HOURS AT RATED POWER OUTPUT

Two new 3-megawatt, S-band Amplitrons have demonstrated an operating life of more than 1,000 hours at rated power output. The QK-622 covers the 2,900 to 3,100 Mc band; the QK-783, the 2,700 to 2,900 Mc band. Both tubes supply full power with low phase pushing characteristics over their entire operating bands at efficiencies greater than 70%--making them unquestionably the most highly efficient microwave tubes thus far developed.

Tubes may be operated at reduced peak power levels to serve as driver stages. High efficiencies are retained at peak power of 600 Kw and gain of 10 db.

Exceptionally long tube life is made possible by the fact that no cathode warmup is required. Starting takes place whenever RF input is present prior to application of modulating pulse. Heater supplies may be omitted entirely from the equipment.

Applications include power-amplifier stages for long-range radars. The tube has been used successfully as an RF power source for linear accelerators.



Excellence in Electronics





Typical Operating Characteristics (QK622 and QK783 Amplitrons)

Peak Power Output (min.)	3 Mw
Average Power Output	15 Kw
Pulse Duration	10 µ sec
Band Width	200 Mc
Duty Cycle	.005
Pulse Voltage	50-55 Kv
Peak Anode Current	65 amps
Efficiency	70%
RF Input	475 Kw
Weight (with permanent magnet)	125 lbs.



You can obtain detailed application information and special development services by contacting: Microwave and Power Tube Division, Raytheon Company, Waltham 54, Massachusetts

A LEADER IN CREATIVE MICROWAVE TECHNOLOGY





(Continued from page 58.4)

and for educational and training programs for marketing personnel.

A native of Vancouver, Wash., he received the B.S. degree in electrical engineering from the University of Washington in 1940. Shortly thereafter, he joined General Electric's Test Program at Schenectady, N. Y., as a student engineer. He subsequently served in various engineering and sales assignments throughout the Company and in 1956 was appointed Pacific regional manager for the Computer Department. He has also worked as a sales engineer for William Miller Instrument Company, manager of field sales for AMPEX Instrument Division, and manager of marketing for AMPEX International.

Mr. Hogg is a member of the Instrument Society of America, Tau Beta Pi, and Sigma Xi.

<mark></mark>*

Dr. Robert C. Hansen (S'47-A'49-M'55-SM'56) has been appointed Senior Staff Engineer in the Telecommunications

Laboratory of the Space Technology Laboratories of Los Angeles, Calif. Prior to this, he was a Senior Staff Engineer in the Hughes Microwave Laboratory.

He obtained the B.S. degree from the Missouri School of Mines and Metallurgy and the



R. C. HANSEN

Ph.D. degree from the University of Illinois. He has been active in IRE affairs, having been PG coordinator for the Los Angeles Section and Chairman of the WESCON Technical Program Committee. He is currently a member of the PGAP and PGMTT Administrative Committees, and is Western Vice Chairman of the Professional Groups Committee.

Dr. Hansen is a member of the American Physical Society, Tau Beta Pi, Sigma Xi, Eta Kappa Nu, and Phi Kappa Phi.

•••

Philip S. Hessinger (SM'39) has been named Manager of Research of National Beryllia Corporation, North Bergen, N. J., according to an

according to an announcement by Christian E. Nelson, President. In the newly-created position, he will be associated with Dr. Eugene Ryshkewich, National Be ryllia's Director of Research, in administration of the company's expanded program of re-



P. S. HESSINGER

search and development relating to be-(Continued on page 64A)



High voltages at high power. Tricky business — but a specialty of the Carad Corporation. Take this 30 kv, 3.3 ampere DC power supply, for instance. Carad designed and built this one in 70 days for Eitel-McCullough, Inc. It supplies beam voltage for production testing of high-power klystrons — 80 to 100 hours a week. One important characteristic is its ability to withstand severe load arcing — and protect the klystron being tested. This Carad supply will clear itself in 30 to 50 milliseconds and includes special reactors to limit current surges. For custom systems involving high-voltage supplies, pulsers, modulators or special transformers, investigate Carad's unusual capabilities.

CARAD CORPORATION

2850 Bay Road • Redwood City • California

DESIGNERS AND MANUFACTURERS OF Systems and pulse components



Engineering progress at Chassis-Trak, keeping pace with the equipment mounting needs of the electronics industry, has resulted in two new slide designs. They are:

1¾-inch slide

Ideal for light-duty slide applications--loads up to 50 lbs. Chassis-Trak "pencil thin" design plus an overall height of only 1.687" saves cabinet space, permits easy mounting without cabinet modification. Cadmium-plated cold-rolled steel construction. Phenol epoxy coating provides permanent dry lubrication. Tilt and non-tilt styles in eight standard lengths---10, 12, 14, 16, 18, 20, 22 and 24 inches.

Lightweight slide

Newly developed model for special equipment mounting problems. Exceptionally compact (1" high, $\frac{1}{2}$ " wide), yet supports up to 150-lb. loads. Saves space without sacrificing heavy-duty strength. Low in cost, easy to install. All stainless steel construction. Precision roller and ball bearings for effortless operation.

Check with Chassis-Trak engineers for the solution to your rack or cabinet application. Slides available in tilt, non-tilt, and tilt-lock models. Supports up to 275 lbs.



For further information contact:



(Continued from page 62.4)

ryllium oxide and other pure metal oxide ceramics.

He was formerly employed at the Mycalex Corporation of America, where he was Acting Director, Research and Development. He has also engaged in and directed ceramics research at Ohio State University Research Foundation and Wright Air Development Center.

Mr. Hessinger is a member of the American Ceramic Society, the National Institute of Ceramic Engineers, and the American Society for Testing Materials. He is the author of many papers and articles on ceramics, and is chairman of the ASTM group on measurement of ceramic insulation value at elevated temperatures. He holds the B.S. and M.S. degrees in Ceramic Engineering from Alfred University and Ohio State University.

•*•

Dr. Harold K. Hughes (A'36 -M'16) has been appointed director of the Department of Physics at the Central Research

and Engineering Division of Continental Can Company in Chicago, it was announced by Curtis E. Maier, general manager of the Division.

He will head the company's research work on application of the principles of physics to high speed automatic



II. K. HUGHES

equipment for the production and quality control of metal, paper, plastic, glass and composite containers and closures, and for any future radiation program. His experience is in the fields of physics and instrumentation and includes work in electronics, radio frequency spectroscopy, molecular beams, applied spectrochemical analysis, X-ray diffraction analysis and statistics.

From 1945 to 1958, Dr. Hughes held positions in the fields of applied physics with the Celanese Corporation of America, Socony-Mobil Oil Co. and the Markite Corporation. During the years 1935 to 1945, he was in the academic world as assistant professor and head of the Physics Department of the University of Newark, instructor in Physics at Columbia University and scientist at the Columbia Radiation Laboratory.

Dr. Hughes received the B.A., M.S., and Ph.D. degrees from Columbia University. He is a member of the American Physical Society, the American Chemical Society, the American Society for Quality Control, the Society of Applied Spectroscopy, the Electro chemical Society and other professional groups.

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(Continued on page 66A)

525 S. Webster, Indianapolis 19, Indiana

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

ONE OF MANY EXAMPLES OF EXCELLENCE IN THE DATA HANDLING FIELD







SOLID STATE ULTRA RELIABLE HIGH PERFORMANCE MILITARY QUALITY

The Data-Stor Model 59 Digital Tape Transport is ideally suited for use in computer, instrumentation and control applications. It incorporates the highly reliable features of military tape transports developed by Cook Electric Company during the past 12 years, and has been proven in the Atlas, Titan, Polaris and other missile programs.

These features include exclusive use of modern ultra reliable solid state circuitry, eliminating gas or vacuum tubes. Precise tape handling is insured by proportional reel drive servo systems that have no jerky step servos. Tension error sensing is accomplished by synchro transmitters with no unreliable potentiometers or contact pile-ups. Field adjustments are eliminated by building tolerances into a single rugged tape deck casting. Endurance and quality are assured by strict adherence to the exacting design and workmanship requirements of MIL-E-4158.

TAPE SPEEDS TO 150 IPS • LESS THAN 3 MS STOP/START • REWIND SPEEDS TO 400 IPS • NO PROGRAMMING RESTRIC-TIONS • PACKING DENSITIES TO 600 NRZ BPI • OPERATES FROM 5 VOLT CONTROL PULSES OR LEVELS OF EITHER POLARITY • FRONT PANEL ACCESS • CHOICE OF NARTB, IBM, OR SPECIAL REELS • ANY TAPE TO 1" • CONDUCTIVE LEADER, LIGHT TRANSMISSIVE, OR LIGHT REFLECTIVE END OF TAPE SENSORS • SOLID STATE READ/WRITE AMPLIFIERS • METAL FACED READ/WRITE MAGNETIC HEADS • AVAILABLE AS HIGH SPEED PHOTOELECTRIC READER.

> Experienced recording systems engineers are invited to apply for existing employment opportunities.



Designers and makers of ground and airborne magnetic recording systems, photoelectric readers and computer peripheral equipments.

ADDRESS YOUR INQUIRIES TO 8100 MONTICELLO AVENUE SKOKIE, ILLINOIS

PROCEEDINGS OF THE IRE March, 1960

Part Number Total V.A. Output O.C. Output F.W. Bridge Volts C.T. Full Wave Ma. Wave Ma. M8034 125 500 250 420 M8035 125 500 250 420	All items	A A A A A A A A A A A A A A A A A A A	ATU SIS SF(DEL DEL DEL DEL DEL DEL DEL DEL DEL DEL			S T T T T T T T T T T T T T T T T T T T
Part Number Total V.A. Output O.C. F. W. Bridge Volts O.C. Ma. Output C.T. Full Wave Volts Wave Ma. M8034 125 500 250 250 420 M8035 125 500 250 250 420	ROBOTRA		Түріс	CIRCL		ER
M8034 125 500 250 250 420 M8035 125 500 250 250 420	Part Number	Total V.A. Output	F. W. I Volts	O.C. C Bridge Ma,	Dutput C.T. Ful Volts	l Wave Ma,
M8036 40 450 90 225 155 M8037 22.5 250 90 125 155	M8034 M8035 M8036 M8037	125	500 500 450	250 250 90	250 250 225	420 420

G	Available in 4 Hermetic (·H) Open Frame (-F wgt. 4 oz.	Case types 15/16" x 11/16 17/16" x 19/16	TURE R '', wt.34 oz. /32'' x 34'',
Part Number	Application	Pri. Imp.	Sec. Imp.
MMT 5*	Coll, to Speaker	50,000	6
MMT 7*	Coll. to P.P. Emit.	25,000	1,200 C.T.
MMT 9*	Line to P.P. Emit	600 C.T	1.200 C.T.
MMT 10*	Coll. to Emit.	25,000	600
MMT 11*	P.P. Coll. to Emit or Li	ne 4,000 C.T	. 600 C.T.
MMT 12*	Coll. to Speaker	2,000	3.4
MMT 16*	Coll. to P.P. Emit.	10,000	1,500 C.T.
MMT 17*	P.P. Coll. to P.P. Emit.	10,000 C.T	200 C.T.
MMT 18*	P.P. Coll. to P.P. Emit.	25,000 C.T.	1,200 C.T.
MMT 19*	(oll. to P.P. Emit.	2,500	2,500 C.T.
"Add eithi See catalo	er -M or -H to part numl og for detailed information	ber to designate	construction



Write TODAY for totalog and price list of the complete MICROTRAN line. MICROTRAN company, inc. 145 E. Mineola Ave., Valley Stream, N.Y.



(Continued from page 64A)

Raymond F. Guy (A'25–M'31–F'39) Senior Staff Engineer of the National Broadcasting Company, has been elected President of the De Forest Pioncers.

He is a pioneer in radio, television and short wave broadcasting. He was a combined announcer and engineer and a wellknown air personality in the earliest days of broadcasting in the New York area. For nearly 30 years he was responsible for planning and construction of all NBC transmitting facilities, which included a leading part in the creation of the pioneering Empire State Building TV tower which

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in Electronics, Automation and

Industrial Electronics Engineering Technology

1.

is shared by all New York stations.

Mr. Guy is a Past President of the IRE, Fellow of the American Institute of Electrical Engineers, a Past President of the Broadcast Pioneers, and First Vice President of the Veteran Wireless Operators Association, an organization of prominent industry veterans of the very early days of wireless. He is Chairman of the Engineering Committee of the Voice of America, for many years was Chairman of the Engineering Committee of the Television Broadcasting Association and the Engineering Advisory Committee of the National Association of Broadcasters, and is active in many other organizations, several of which have honored him with medals of achievement and special citations

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(Continued on page 68.4)



Request your free Home Study or Resident School Catalog by criting to: Dept. 2603G 3224 16th St., N. W. Washington 10, D. C. Approved for Veteran Training During IRE Show, visit our booth No. 4430



CUSTOM MADE TO MEET ANY REQUIREMENT

STAMPED VINYL PLASTIC SLEEVING LAMINATED SNAP-ON SPLIT SLEEVES STAMPED SILICONE RUBBER SLEEVING

Permatag wire markers are furnished in bulk or carded in sequence according to blue prints or wiring diagrams. Phone or write for samples and quotations.

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RESINITE SOFT-WOUND INSULATION SLEEVING IS AVAILABLE FROM STOCK

• "Soft-Wound" is a new concept in spooling and coiling insulation sleeving. It is carried in stock to meet the following mil specs: MIL-I-631C, MIL-I-7444A(2), MIL-I-3190.

"AIR-TEX" Electronic Lacing Cords to Mil-T-713A Mil-C-5649B

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P12003



WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960



*RELIABILITY A Rare Combination: *FLEXIBILITY *SERVICEABILITY

CHICAGO DYNANIC INDUSTRIES, PRECISION PRODUCTS DIVISION CHICAGO, ILLINOIS

PRESENTLY SUPPLIED FOR CRITICAL MILITARY APPLICATIONS

Ideal for any application requiring high reliability. Can be furnished to meet applicable requirements of MIL-S-3786 and MIL-E-5272.

Unusual Flexibility

- Quick program changing.
- Quick configuration changing.
- Quick circuit changing.

-C·D·I)-

5-SECOND WAFER REPLACEMENT

Here's all you do:

Turn Dial to Top Position and withdraw dust cover,

Snap out old wafer, snap in replacement wafer, restore dust cover.

Servicing is finished in seconds. No time-wasting disassembling, wire removing or soldering. No skilled technicians needed for wafer changing.

OTHER FEATURES

- All connections on one side for easy access and wiring.
- Lower torque than most standard switches.

Manually, motor or solenoid operated rotary switches are available in sizes approx. $2^{\prime\prime} \times 2^{\prime\prime}$, $3^{\prime\prime} \times 3^{\prime\prime}$ and $4^{\prime\prime} \times 4^{\prime\prime}$ with lengths to accommodate up to 36 wafers. Virtually unlimited choice of switch circuit configurations. Manufactured under U. S. Patent No. 2,841,660. Other U. S. and foreign patents pending. Write today for technical literature.

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Driver Rodar is processed from melting to finished size in our own plant under the strictest controls to insure consistent analysis, temper, uniform grain size and conformance to customers' specifications. The superior stamping and sealing properties of Rodar make it the preferred sealing alloy.

Rodar produces a permanent, vacuumtight seal with simple oxidation procedure and resists attack by mercury. Readily machined and fabricated, Rodar can be welded, soldered or brazed. Available in wire, strip and bar to your specifications.

Average Thermal Expansion, *Cm/Cm/°Cx10^{.6} Temperature Range VISIT BOOTHS 4.33 To 5.30 30° To 200 C. 4201-30° To 300 C. 4.41 Te 5.17 4203 4.54 Te 5.08 30° Te 400 C. IRE 30° To 450 C. 5.03 To 5.37 SHOW 5.71 To 6.21 30° Te 500 C. *As determined from cooling curves, after annealing in hydrogen for one hour at 900°C, and for 15 minutes at 1100°C. WILBUR B. DRIVER CO. NEWARK 4, NEW JERSEY

IN CANADA: Canadian Wilbur B. Driver Company, Ltd., 50 Ronson Drive, Rexdale (Toronto)

IRE People

(Continued from page 66.4)

Dr. Robert R. Johnson (S'50-M'56) will be responsible for engineering design and development of the General Electric

Company's Computer Department's complete line of computers and computing systems. A native of Madison. Wis., he joined General Electric in 1950 as à test engineer Schenectady, at Y., following N. graduation from the University of Wisconsin with the



R. R. JOHNSON

B.S. degree in Electrical Engineering, He received the Master's degree from Yale University in 1951, specializing in servomechanisms, and in 1955 was awarded the Ph.D. degree in electrical engineering at California Institute of Technology

Dr. Johnson has been awarded several patents and has written chapters for two computer handbooks. He is a member of the American Institute of Electrical Engineers ÷.,

David Y. Keim (A'36-VA'39-M'55) has been appointed director of engineering in Stromberg-Carlson's Electronics Division has been an-

nounced by Kenneth M. Lord, vice president and general manager. Stromberg-Carlson is a division of General Dynamics Corporation.

In his new position he will be responsible for the administration of design and develop-



ment engineering work in the division.

He joined Stromberg-Carlson in March, 1959, as chief engineer of military products in the Electronics Division. He previously served as engineering department head for microwave and electronic equipment and also directed advanced research work in the field of microwave devices for the Sperry Gyroscope Company, Earlier he was employed by Sylvania Products Company

Mr. Keim received the B.S.E.E. degree from Pennsylvania State University in 1936. He is a member of the American Institute of Electrical Engineers and the American Ordnance Association. He has contributed a number of papers on weapons support equipment and related subjects to technical journals.

Dr. Robert C. Langford $(\rm M'54\text{-}SM'55)$ has been appointed director of engineering of the Newark Operations of Weston Instruments Division of Daystrom, Incorpo-

(Continued on page 72.4)



WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

... 1450°C. (Approx.)

.... 294 Ohms C.M.F.

..... 80,000 PSI

.... 82 B Rockwell

30% (2" gauge length)

Specific Gravity . . 8.36

Weight Per Cubic Inch

Electrical Resistivity

Tensile Strength

Hardness

Elongation

Melting Point

March, 1960

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... help you safeguard your product's reputation for Quality and Reliability!

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That's why it doesn't pay to gamble with fuses that could be faulty and create trouble for your customers either by failing to protect and causing useless damage to equipment, or by blowing needlessly and causing unnecessary shutdowns. With BUSS and FUSETRON fuses safe, dependable electrical protection is assured. Before one of these fuses ever leaves our plant, it is electronically tested to make sure it is right in every way... to make sure it will protect, not blow needlessly.

When you specify BUSS or FUSE-TRON fuses, you are safeguarding against customer complaints for you have equipped your product with the finest electrical protection possible. You are also helping to maintain the reputation of your product for service and reliability. To meet all fuse requirements, there's a complete line of BUSS and FUSETRON fuses in all sizes and types...plus a companion line of fuse clips, blocks and holders.

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759





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Here is Francis Alterman, Manager of General Mills Digital Computer Laboratory, checking one of our newest computers which he helped design. General Mills computers, both analog and digital, are being used in missile

guidance, bombing and navigation systems, automatic surveying and in industrial control. In future space travel, computers will help control navigational systems of space vehicles and will process data gathered in outer space.

General Mills engineers work today

General Mills has been producing computers for nearly 20 years. Exciting new concepts in high speed magnetic tape units, ultra-high precision analog to digital converters and optical keyboards are examples of continuous developments in our over-all computer program. We work to improve reliability, increase speed, cut cost.

Our research activities cover broad areas in physics, chemistry, mechanics, electronics

and mathematics. Some of the studies representative of these activities are: ions in vacuum, deuterium sputtering, dust erosion, magnetic materials, stress measurements, surface friction and phenomena, trajectory data and infrared surveillance.

In our engineering department, current projects include: specialized inflatable vehicles and structures, airborne early warning systems, micro wave radar test equipment,


Mars seen from one of its moons . . . illustration from book written for General Mills by Willy Ley.

. to help you explore space tomorrow

antennas and pedestals, infrared and optics, inertial guidance and navigation, digital computers—and many other activities.

Our manufacturing department is geared to produce systems, sub-systems, and assemblies to stringent military requirements. STOP AT GENERAL MILLS EXHIBIT BOOTHS 3937, 3939

IRE SHOW

New Concepts in Computers

New York Coliseum

March 21-24, 1960

General

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To wider worlds-through Intensive Research • Creative Engineering • Precision Manufacturing



NEW FROM NARDA



MICROWAVE

accepts over 40 magnetrons!

Here's the first of a series of new products from Narda's recently-established High Power Electronics Division! A high power Microwave Modulator that permits installation inside the unit of any of more than 40 magnetrons! Complete, compact and self-contained, it accepts magnetrons covering 3,200 mc to 35,000 mc, with peak outputs from 6 KW to 120 KW. Model 10001 features a completely interlocked circuit, with all high voltage leads and connections internal, for maximum safety; solid state high voltage bridge rectifiers for longer life and reduced heat output (prolonging life of other components, too); and built-in meters and viewing connectors for all principal parameters.

Other features are shown below. For complete specs and a list of at least 40 magnetrons suitable for use with the 10001, write Narda's High Power Electronics Division (HPED) at Dept. PIRE-7.

SPECIFICATIONS

High voltage supply: Continuously variable from 0 to 4 KV at 100 ma; Pulse power: 18 KV at 20 amps max.; Magnetron filament supply: Cont. variable from 0 to 13 volts at 3 A; Rep. rate generator range: Cont. variable from 180 to 3000 pps; Pulse width: 1 microsecond at 70% points, rise time 0.15 microseconds, max. slope 5% (other pulse widths available); Size: 38" h, 22" w, 18" d. Weight: 150 lbs.

Complete 1959 catalog available on request.



IRE People /

(Continued from page 68.4)

rated. The plant is a unit in Daystrom's Industrial Products Group.

He succeeds Francis X. Lamb (A'36– VA'39–M'55), formerly the Newark plant's vice president of engineering, and for 30 years an outstanding figure in the development of techniques and devices for the measurement of electricity, Mr. Lamb was named engineering consultant to J. F. Degen, vice president of Operations, at Newark.

Both appointments, effective immediately, were announced by Mr. Degen, who said that Dr. Langford will be responsible for Weston-Newark's Engineering Services and Product Development.

Dr. Langford was chief engineer for Research and Development at the Newark Operation.

A native of Portsmouth, England, he obtained the B.S. degree (cum lande) in 1944 from the University of London, which he attended on the Sylvania-Thompson Scholarship awarded him by the Institute of Electrical Engineers. He was subsequently granted the Swan Research Fellowship and attained his doctorate from Queen Mary College, London.

He is a holder of several patents on instrumentation and author of many papers, including co-authorship on the subject in standard technical reference books.

Dr. Langford is a member of the American Institute of Electrical Engineers and the American Nuclear Society. He is chairman of the Instrumentation Division of the AIEE in New York. He is also chairman of the IRE Professional Group on Engineering Management of the IRE and a member of the Technical Program Committee planning the program content of the 1960 IRE Convention.

In a move to expand the company's operations in advanced military and industrial electronics, Packard Bell Elec-

tronics Corp. has announced the formation of a new Defense and Industry Gronp, encompassing two divisions and a subsidiary.

Affected are the Technical Products Division; the Packard Bell Computer Corporation, which has been operating

as a division; and Technical Industries Corporation of Pasadena, a subsidiary.

Robert S. Bell, president, said **Richard B. Leng** (SM'53) former vice president in charge of the Technical Products Division, will head the new organization as group vice president to coordinate more closely the design, engineering, development, manufacturing and marketing of electronic instruments and systems.

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(Centinued on page 74.4).



R. B. LENG.

^{••}

SIX VSWR AMPLIFIER FEATURESavailable only from NARDA

- 1. Battery-operated (rechargeable nickel-cadmium).
- 2. Completely transistorized for low current drain.
- 3. Independent of line voltage variations.
- 4. Complete bolometer protection during switching.
- 5. Most compact unit available.
- 6. Completely portable.

Model 441B-\$225

Now you can get a completely portable batteryoperated VSWR Amplifier offering complete protection against bolometer burnout at the same time!

Narda's Model 441B is supplied with nickel-cadmium batteries, providing complete freedom from line voltage deviations. Batteries recharge automatically when unit is plugged in; provision is built-in to show state of battery charge. A special protective circuit

FEATURES:

- SENSITIVITY: 0.1 microvolts at 200 ohms for full scale.
- FREQUENCY: 1,000 cps \pm 1% (plug-in frequency networks available for 315-4,000 cps and broad-band applications)
- BANDWIDTH: 25-30 cps
- RANGE: 72 db (60 db in 10 db steps, 11 db continuous)
- ACCURACY: ± 0.1 db per step ± 0.2 db maximum cumulative meter linearity: 1% of full scale





danger of bolometer burnout. Provision is made for both crystals and high and low current bolometers. Full sensitivity is provided over both normal and expanded scales: eliminates switching attenue time

permits switching and connect-disconnect with no

expanded scales; eliminates switching attenuation range. Other features are shown on this page; for complete information and a free copy of our latest catalog, write to us at: Department PIRE-10.







STANDARD EQUIPMENT

Two face plates One collet draw-in bar One 6-fire seven jet burner assembly Hand carburetion control Foot pedal control of air or nitrogen supply and of oxygen-gas volume





Main air valve controlling air in either or both spindles 1/2 h.p. Motor, 230 volt, three phase, single speed, 60 cycle, AC Face plate wrench One motor belt

One motor pulley

Radial clearance abave apran .

General Specifications

Variable Speed Drive - Electronic Control (As shown) Available at extra cast Spindle hale diameter

Maximum width averall . 24" Maximum length spindle nase . . 47" ta spindle nase

Net weight Apprax, 1250 paunds Approx. shipping weight . . . 1500 pounds

Litton Engineering Laboratories

Grass Valley, California · P. O. Box 949





(Continued trem page 72.1)

Election of John J. McDonald (N'50) as a vice president of Consolidated Systems Corp., a wholly owned subsidiary of Consolidated Electro-

dynamics Corp., has been announced by Kennett W. Patrick. CSC president

He was appointed director of engineering when CSC was incorporated as a subsidiary of Consolidated Electrodynamics on March 1



McDos MD

of this year. He will continue at this post. Previously, he was assistant director of the Systems Division for three years and manager of CEC's Central Regional Office in Chicago for five years.

Mr. McDonald is a vice president of the Instrument Society of America and has served the ISA as a national director, director of the Transportation Industry Division, and chairman of the Research and Development Committee. He is a member of the American Institute of Electrical Engineers, Society for Experimental Stress Analysis, American Standards Association, and American Rocket Society. He received the B.S. degree in physics from the University of Chicago and the B.S. degree in electrical engineering from the Armour Institute of Technology.

2.

Dr. John C. McGregor (M'46-SM'49), president of The Narda Ultrasonics Corporation, Westbury, N. Y., has taken on the added responsibil-

ity of marketing of the wares of the firm.

Physicist. and lawyer, Dr. Mealso Gregor 15 chairman of the board of the related Narda Microwave Corporation, developer of electronic items. He also is board chairman



J. C. MCGREGOR

of Technical Information Corporation and of Harper and Saladino, Inc., industrial art specialists, and a director of North Atlantic Industries.

Dr. McGregor received the B.S. degree in physics from Carnegie Institute of Technology in 1941; the LL.B. from St. John's University (New York) in 1947, and the Doctorate of Jurisprudence from Brooklyn Law School in 1949. He is a member of Sigma Nu, Phi Delta Phi (legal), the Nassau County Bar Association, and the Institute of Aeronautic Sciences. He is admitted to practice in the New York State, Federal and Patent Courts.

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(Continued on page 70.4)

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World Radio History

71/4

13 3/1

5173

THE RECORDING THAT WASN'T

... It's happened to lots of magnetic tape users



Test factually demonstrates shielding effectiveness of Netic alloy material and enclosure design. Instrumentation used: magnetic field radiating source. AC vacuum tube voltmeter, Variac, pickup probe and Netic Tape Data Preserver. For complete test details and results, request Data Sheet 142.



For safe, distortion-free storage of large quantities of vital magnetic tapes. Designed for Military Establishments, Radio & TV Broadcasters, Automated Plants, Libraries, Laboratories, Gov't. Agencies, etc.



sties of NETIC alloy material are not affected by vibration, shock (including dropping) etc. Furthermore, NETIC does not retain residual magnetism nor require periodic annealing. Maybe you've been one of these unfortunates . . . who've spent thousands of dollars . . . plus many man hours . . . to record valuable information on magnetic tapes . . . only to find the data useless from accidental distortion or erasure.

Unexpected exposure to an unpredicted magnetic field, and presto!—your valuable data is filled with irritating odd noises. Distortions may result in virtual data erasure.

Unprepared tape users never realize the danger of loss until it's too late.

Such losses have become increasingly common from damaging magnetic fields during transportation or storage. These fields may be produced by airplane radar or generating equipment or other power accessories. Also by generators, power lines, power supplies, motors, transformers, welding machines, magnetic tables on surface grinders, magnetic chucks, degaussers, solenoids, etc.

Since 1956, many military and commercial tape users successfully avoid such unpleasant surprises. Their solution is shipping and storing valuable tapes in sturdy NETIC Tape Data Preservers.

Data remains clear, distinct and distortion-free in NETIC Preservers. Original recorded fidelity is permanently maintained.

Don't take chances with your valuable magnetic tapes. Keep them *permanently clear* and *distinct* for *every* year of their useful life in dependable NETIC Preservers. Can be supplied in virtually any size and shape to your requirement. Write for further details today.



For complete, distortion-free protection of valuable tapes during transportation or storage. Single or multiple containers available in many convenient sizes or shapes.

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EYELETS, RIVETS, GROMMETS, WASHERS, HOLE PLUGS SNAP FASTENERS, FERRULES, TERMINALS, STAMPINGS

and many similar fasteners are made in enormous variety and quantity. Made from most any metal and in all finishes. We also make a complete line of machines for attaching eyelets, rivets, etc.

Send for our general catalog which illustrates over 1000 metal articles.





(Continued from page 74.4)

John S. McCullough (A'47-M'51-SM'56) has joined Litton Industries Electron Tube Division, San Carlos, Calif., as assistant to the

general manager. He will be responsible for new product planning, according to Dr. Norman II Moore, Litton vice president and division general manager.

Previously director of research and engineering of Eitel-McCullough,



I. S. MCCULLOUGH

Inc., Mr. McCullough attended the University of California and Harvard University.

Daniel G. O'Connor (A'46–M'48– SM'52) has been appointed to the position of Assistant to Dr. R. L. Garman, Vice President in charge of engineering and research of the General Precision Equipment Corp. Ho will evaluate proposals and programs involving digital computers, and has been named a member of the Engineering Plans Committee,

He was transferred from the Link Di-

(Continued on page 78-4.)





with new Hughes "20-20" Circulators!

With 20% bandwidth and over 20 db isolation, the new Hughes "Y" and "T" Circulators are ideally suited for microwave reception and transmission applications. They also give you small size and weight...without sacrifice in performance. C- and X-Band models are available today!

For information on the new "20-20" Circulators, or other advanced microwave components, please write Microwave Products Department, Advanced Program Development, Hughes Aircraft Company, Culver City 6, California. Or, phone UPton 0-7111, Ext. 6919.

	Model C-201A	Model X-230A (Illustrated)	
Frequency:	4.9-6.2 Kmc	8.0-9.8 Kmc	
Isolation	20 db	20 db	
Insertion Loss:	0.3 db	0.3 db	
Input VSWR:	1.10	1.20	
Power Canacity	10 Kw nesk	3 Kw neek	

Power Capacity. TU Kw peak 3 Kw peak 100W avg (Min.) 50W avg (Min.) **ALSO AVAILABLE:** Miniaturized Sand L- Band Coaxial Circulators. New, extremely small $(1'' \times 2'' \times 8'')$ circulators with bandwidths to 10%, over 20 db isolation, and 0.5 db insertion loss are now available.

Creating a new world with ELECTRONICS



PROCEEDINGS OF THE IRE March, 1960



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in electrical and

control systems

LINEAR CIRCUITS By Ronald E. Scott

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An introduction to linear circuits for students of electrical engineering. Assumes an introductory course in physics and concurrent calculus. Noteworthy features include: a balanced treatment of time and frequency domain methods; an integrated treatment of Fourier series and integrals, Laplace transforms, and power density spectra; and coverage of tapics seldam faund in backs at this level, such as signal flow graphs, relaxation methods, dummy voriables, s-plane plots, and complex resonance.

> C. 400 pp., 1,000 illus., to be published Summer 1960—price to be onnounced

ANALYSIS OF LINEAR SYSTEMS

By David K. Cheng

Syracuse University

"A very fine book for the personal library of every engineer who has need for ready reference on linear systems theory and the use of Loplace and Z transforms."

> Proceedings of the IRE 431 pp., 265 illus., 1959—\$8.50

ORDINARY DIFFERENTIAL EQUATIONS

By Wilfred Kaplan

University of Michigan

Deals thoroughly with all standard methods of integration . . . input-output analysis is given a prominent place . . . a thorough study is made . . . by means of phose-plane analysis . . . of great value to practicing engineers."

> Applied Mechanics Reviews 534 pp., 150 illus., 1958—\$9.75

ENGINEERING SYSTEMS ANALYSIS

By Robert L. Sutherland

State University of Iowa

"The second-order lineor differential equation ... is treated in detail for mechanical, electrical, and occustical systems ... offers excellent introductions to dimensional analysis, feedback and control, and analog and digital computers."

> Physics Todoy 223 pp., 98 illus., 1958—\$7.50







(Continued from page 76.4)

vision of General Precision, Inc., a subsidiary of General Precision Equipment Corp., where he was responsible for a number of special projects including the Link Target Programmer, the Link Digital Function Generator, DOTitron and other special purpose digital computers. General Precision, Inc., provides all major military services, business and industry, with advanced electronic control and support systems.

Mr. O'Connor's experience includes four years of engineering planning and development activities in data processing at IBM, and five years research and development at the Physics Laboratory of Sylvania Electric Products, where he worked on guidance systems and electronic devices representing new advances in information theory. He has taught at IBM schools and in the Graduate School of Electrical Engineering at Polytechnic Institute of Brooklyn. He has done graduate work at Syracuse and Stauford Universities and the Polytechnic Institute of Brooklyn.

**

Donald E. Nasoni (A'54) has been appointed staff engineer in Advanced Systems Simulation at the Owego facility of IBM's Federal Systems Division.

He entered Systems Studies at Owego in November, 1958 as an associate engineer. He was assigned to his present department last June.

A native of Scranton, Pa., he graduated from Central high school there and earned the B.S. degree in Electrical Engineering from Pennsylvania State University in 1952.

Mr. Nasoni is a member of the Association for Computing Machinery and the Scientific Research Society of America.

¢.

Charles A. Parry (SM'53) international authority on telecommunication systems planning, has been named to head the newly-estab-

lished Telecommunications Directorateat Page Communications Engineers, Inc., a Northrop Corporation subsidiary. Since coming to Page from RCA International in 1957, he has been consultant on the firm's overseas telecommuni-



C. A. PARRY

cations projects. He will continue as Vice President-Engineering for the company's Italian affiliate, Edison-Page.

···

(Centinued on page 80.4)









Now in production by Bendix* are eight 25-ampere peak current power transistors capable of switching up to 1000 watts—and you can get immediate delivery on all eight types.

NOW!

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POWER TRANSISTOR

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25-AN

Newly improved in design, the transistors have a higher gain and flatter beta curve. The series is categorized in gain and voltage breakdown to provide optimum matching and to eliminate burn-out.

Current Cain	Maximum Voltage Rating			
hFE at $Ic = 10$ Adc	50 Vcb	60 Vcb	90 Vcb	100 Vcb
	30 Vce	40 Vce	70 Vce	80 Vce
20-60	2N1031	2N1031A	2N1031B	2N1031C
50-100	2N1032	2N1032A	2N1032B	2N1032C

Ask for complete details on this newly improved Bendix transistor series . . . and on the entire Bendix line of power transistors and power rectifiers. Write SEMICONDUCTOR PROD-UCTS, BENDIX AVIATION CORPORATION, LONG BRANCH, NEW JERSEY, or the nearest sales office.

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Microwave energy normally travels on a two-way street. Because power flow has the

habit of being reciprocal, it often becomes vitally necessary for microwave systems

engineers to isolate the load from the transmitter or to control power channeling.



Rantec LOAD ISOLATORS directionalize the

power flow with a minimum loss in one direction of propagation and with extremely

high loss in the other. Rantec manufactures a



large number of "off-the-shelf" LOAD ISOLATORS which range from low power to

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Rantec engineers welcome the opportunity of solving specific problems or undertaking

special design projects.





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microwave ferrite components / antennas / r-f devices SEE RANTEC AT PLAZA HOTEL DURING THE IRE SHOW

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



(Centinned from page 78.4)

The appointment of **C. Robert Paulson** (S'48-A'50-M'52) to manager of the Professional Audio Products Division of Am-

pex Professional Products Company has been announced by Neal K. Mc-Naughten, Ampex Corporation vice president and manager of the Ampex Professional Products Company, a totally integrated division of Ampex Corporation.



C. R. PAULSON

Mr. Paulson re-

places Frank G. Lennert who will remain with the company in an advisory capacity on audio matters.

As well as taking on his new duties as division manager, Mr. Paulson will continue in the position of marketing manager for the Professional Audio Products Division until a replacement is appointed.

He started with Ampex as New York district audio sales manager in 1953 and later moved to the main Redwood City offices as the first National sales manager for the Professional Products Division.

Prior to joining Ampex, he was sales engineer for Audio-Video Products, New York City; assistant producer for the Fred Waring Television Show; and staff engineer for the Thayer School of Engineering.

A graduate of Dartmouth College, he is a member of the Audio Engineering Society, the Armed Forces Communications System and the San Francisco Sales Executives Association.

During World War H, Mr. Paulson served as First Lieutenant with the Signal Corps, at Camp Crowder, Fort Monmouth and with the Allied Force headquarters in Italy.

•

The Baltimore Division of The Martin Co. is being reorganized to take full advantage of its design, development and production capabil-

ities for Space Age electronic, missile and other weapons systems.

A. L. Varrieur, Vice President and General Manager, said the reorganization is to take effect immediately. It will include the establishment of a

Martin-Baltimore Electronics Division with integrated engi-

neering and production facilities. Heading the Electronics Division as General Manager, reporting to Mr. Varrieur, is **John J.** Slattery (J'29-A'30-SM'46-F'56), former manager of West Coast Operations for the Magnavox Company's Government and Industry Division

(Continued on pure \$2.1)

J. J. SLATTERY

New, Electro Instruments all-electronic. totally transistorized igital voltmeter

50 conversions per second • 1000 megohms input impedance • Fully automatic ranging



Model 8409 Voltmeter and Ratiometer

all the features you want in a medium-speed digital voltmeter

- 3 ranges, 9.999/99.99/999.9 volts
- Automatic, manual and remote ranging
- Automatic polarity
- One digit accuracy
- 4 digit in-line visual readout
- BCD and decimal electrical output
- Direct printer operation—local/remote control

Ask your 📮 representative for complete information.

- New 5¼" x 19" front panel
- Modular construction throughout
- Provision for external reference voltage

Plus accessory modules for every application

AC: All transistorized Model 110; considerably faster AC/DC conver-sion than presently available models. Fully automatic ranging and direct AC voltage readout on the Model 8409.



Ohmmeter: All transistorized Model KIM-000. Provides constant current through test resistor with negligible power dissipation. Voltage measurements made across resistor and read out directly in ohms with fully automatic ranging.

Also scanners, code converter modules, print control modules and many others to solve all digital problems from simple voltmeter applications to complex data logging systems.





NOW . . . THE WORLD'S LARGEST SELLING VTVM in wired or kit form

- ETCHED CIRCUIT BOARDS FOR EASY ASSEMBLY, STABLE PERFORMANCE
- 1% PRECISION RESISTORS FOR HIGH ACCURACY
- LARGE, EASY-TO-READ 41/2" 200 UA METER

The fact that the V-7A has found its way into more shops, labs and homes around the world than any other single instrument of its kind attests to its amazing popularity and proven design. Featured are seven AC (RMS) and DC voltage ranges up to 1500; seven peak-to-peak ranges up to 4,000; and seven ohmmeter ranges with multiplying factors from unity to one million. A zero center scale db range is provided and a convenient polarity reversing switch is employed for DC operation, making it unnecessary to reverse test leads when alternately checking plus and minus voltages.

A large $4\frac{1}{2}$ " meter is used for indication. with clear, sharp calibrations for all ranges. Precision 1% resistors are used for high accuracy and the printed circuit board gives high circuit stability and speeds assembly. The 11-megohm input resistance of the V-7A reduces "loading" of the circuit under test resulting in greater accuracy. Whether you order the factory wired ready-to-use model or the easy-to-assemble kit, you will find the V-7A one of the finest investments you can make in electronic workshop or lab equipment.

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a subsidiary of Daystrom, Inc.	ADDRESSZONESTATE	
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(Continued from page 80.4)

at Los Angeles. His experience also includes design and development assignments with Bell Telephone Laboratories and RCA Manufacturing Company, and both civilian and military duty with the U. S. Army Signal Corps,

At the Signal Corps Engineering Laboratories he helped develop radar and counter-measure systems, and was responsible for systems design and engineering of the first U. S. searchlight control radar. He holds the B.S. degree in electrical engineering from Villanova University and the M.S. degree in physics and mathematics from Stevens Institute of Technology.

B. Linn Soule (M'59) has been appointed liaison engineer for Packard Bell Electronics in the Dayton, Ohio office.

He was employed for the past eight years by the Hazeltine Corporation as a held engineer specializing in systems of ground radar, airborne early warning radar, mission and traffic control, radar judicating and Doppler battlefield surveillance. In ad-



B. L. Soule

dition to assignments in the United States, he has also served in Japan, Korea, Puerto Rico, Alaska and the European continent.

Mr. Soule was graduated from the University of Michigan with the B.S. degree in electrical engineering. He holds an amateur radio operator's license and is a member of the Ohio Society of Professional Engineers and the National Society of Professional Engineers.

*

The appointment of **Robert S. Stein** (A'47–SM'53) as Director of Technical Publication Services has been announced by Boland & Boyce,

Inc., technical publications specialists with offices in New York, N. Y., and Madison, N. J. Mr. Stein will assume full responsibility for technical mamials, brochures and other publications prepared by Boland & Boyce.



R. S. Stets

He joins Boland

& Boyce with a thorough background in the field of technical publications. For the last nine years he was editor-in-chief of Coastal Publications Corp. Prior to that, he held positions as project director, editor, writer and instructor for several other firms in the same field

After receiving his degree in electronic

Centinued on page 814)

March, 1960

what is the frequency standard for the U.S.A.?

ANSWER: By act of Congress, the U.S. Bureau of Standards determines the primary standard, based on the revolution of the earth. All DeMornay-Bonardi microwave instruments are calibrated at frequencies which are verified by our secondary standard, which, in turn, is periodically calibrated, point for point, by the U.S. Bureau of Standards.

One way to properly match a microwave transmission line is by using a D-B Stub Tuner to reduce mismatch losses and utilize the total energy available.

D-B stub tuners in the 2.6 to 18 KMC range have a new scale and vernier that gives precise resettability in longitudinal travel. A new micrometer scale on the probe meas-



ures penetration with very high accuracy.

Probe wobble is eliminated, and no resonances can occur under any conditions. You can correct VSWR as high as 20:1 with amazing accuracy (1.02). You can tune with precision...reset to original settings with certainty that phase and magnitude have been duplicated.

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Write for data sheets—they detail all features, applications, dimensions, sizes. Bulletin DB-919.

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PROCEEDINGS OF THE IRE March, 1960





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JAMES MILLEN MFG. CO., INC.

MAIN OFFICE AND FACTORY MALDEN MASSACHUSETTS





(Centinuel from page 82.1)

engineering from Georgia Institute of Technology in 1944, Mr. Stein served as a radio officer in the Army Signal Corps. Upon completing his military service he did graduate work in television engineering at Columbia University.

÷

The Sperry Microwave Electronics Company recently announced the promotion of **Leonard Swern** (*N*49–M'53–SM'57) to Engineering De-

b rungine Head. In this capacity he is responsible for all microwave solid state activities, as well as research and development of microwave system components and antennas.

A former Sec-

tion Head for Ap-



L. Swern

plied Physics, he joined Sperry Microwave ten years ago after receiving the M.A. degree in physics from Columbia University.

He is a member of the American Physical Society, and is serving on the Editorial Board of the IRE TRANSACTIONS ON MICROWAVE THEORY AND TECHNIQUES. Past works have included design of microwave bolometers, research involving interactions between microwaves and matter, and microwave research and advanced developmental work on the properties of ferrites and other anistropic media at microwave frequencies.

•••

Advancement of two staff members in Stromberg-Carlson's Research Division to new responsibilities has been announced



R. B. TAYLOR

E. G. BROCK

by Dr. Nisson A. Finkelstein, assistant vice president and director of research. Stromberg Carlson is a division of General Dynamics Corporation. The men, and their new positions, are: **Robert B. Taylor** (M'57–SM'59), manager of engineering services, and **Dr. Ernest G. Brock** (SM'56), a principal scientist in the Basic Science Laboratory.

Mr. Taylor has been with Stromberg-Carlson since 1955, when he joined the company as a research engineer specializ-

(Centraned on page 88,1)



Low Noise



VHF and UHF Amplifiers and Preamplifiers SERIES 1000

For application as receiver preamplifiers or wide band i. f. amplifiers . . . in scatter communications systems, laboratory, or nuclear research. Eight standard models cover VHF and UHF to 900 mc. High gain, low noise. Special pass bands available.

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- 120 foot paper length 200 hours recording time.
- Powered by standard "D" cells.
- Fully transistorized --- Completely modular construction

- Available with many accessories; - Plug-in pre-amplifier for 10 mv. full scale sensitivity.
- Transducer assemblies for conversion uses. - Rechargeable batteries module.
- Field applications: Humidity, Temperature, Vibration, Acceleration, etc., recording.

Serviced by Systems Engineering Offices of Airsupply-Aero Engineering Division of the Garrett Corporation Offices in All Principal Cities •

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100 millivolt full scale sensitivity Accuracy — better than $\frac{1}{2}\%$ full scale. Tested for operation in rugged environments.

- Enclosed paper take-up spool. •



Distinctively styled for the '60's, General Electric's BIG LOOK small panel meters offer modern *appearance*, excellent *readability* and improved *reliability*...for *all* your products and equipment. Now including $2\frac{1}{2}$ -, $3\frac{1}{2}$ -, and NEW $4\frac{1}{2}$ -inch a-c and d-c designs, this is the only complete family of fine meters that gives you the built-in advantages of the BIG LOOK. You get a BIG BUY with the BIG LOOK ... here's why: CLEAN-LINE DESIGN not only enhances appearance but also makes possible up to 28 percent increase in scale length over less functional designs ... provides at-a-glance readability by creating the illusion of bigness. Yet BIG LOOK meters actually fit into the same usable panel space as ordinary meters.SEALED CASES on $2\frac{1}{2}$ - and $3\frac{1}{2}$ -inch designs assure complete protection of internal mechanism by locking contaminants out ... reliability in. SNAP-ON, SNAP-OFF COVER is featured on new $4\frac{1}{2}$ -inch design.SELF-SHIELDED D-C MECHANISM, unlike many other designs, puts the core *around* the magnet ... where it belongs. Completely eliminates inter-action, ends special calibration for mounting on magnetic or non-magnetic panels and minimizes



PANEL METERS NOW NEW 4¹/₂-INCH DESIGN

stray field effect. A nationwide network of G-E Apparatus Sales Offices, warehouses and authorized stocking distributors provides on-time delivery. Fast service on specials and prototypes too! And the price is right! CHECK INTO THE ADVANTAGES of the BIG LOOK today. Find out how the BIG LOOK gives you the BIG BUY in small panel meters . . . helps you improve appearance and reliability of your equipment . . . and at low cost! For the full story, contact your local G-E Apparatus Sales Office, or write for bulletin GEA-7034, Section 597-01, General Electric Co., Schenectady, N.Y.

See General Electric's Family of BIG LOOK Small Panel Meters at Booth 2928-2932, I.R.E. Show, New York Coliseum, March 21-24.

INSTRUMENT DEPARTMENT

GENERAL 🍪 ELECTRIC



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(Continued from page 84A)

ing in solid-state circuit design. Subsequently he became staff engineer to the director of research, with principal responsibility for the negotiation and administration of reimbursable research contracts. He received the B.S. degree in electrical engineering from the University of Rochester, and prior to joining Stromberg-Carlson worked as a research engineer at the Delco Appliance Division of General Motors Corporation. He is an officer in the U. S. Naval Research Reserve Company 3-4.

Dr. Brock joined Stromberg-Carlson's Research Division in 1958 as a senior physicist, and has been specializing in studies of molecular resonance. He received the B.S. and Ph.D. degrees from the University of Notre Dame, and before coming with Stromberg-Carlson served as a research associate in the General Electric Research Laboratory, Schenectady, N. Y., and as a group leader at the Linfield Research Institute, McMinnville, Oregon. He is a member of the American Physical Society.

•

Announcement of the appointment of **Raymond W. Wells** (SM'55) as Director of Engineering has been made by Joel M.

Jacobson, President of Aircraft Armaments, Inc., Cockeysville, Md. Mr. Wells will be responsible for the direction of all development and production engineering activities at Aircraft Armaments, Inc., including the management of the company's six engi-



R. W. Wells

neering departments: Electronics, Electro-Mechanical, Structures, Ordnance, Aerodynamics and Nuclear Physics, and Engineering Services. He has been with Aircraft Armaments, Inc., since 1953, when he joined the organization as Project Engineer. He served as Chief Electronics Engineer from 1955 to his present assignment and in this capacity supervised research and development of numerous, advanced electronic programs, including countermeasures, missile and radar test equipment and instrumentation, and microwave and antenna systems.

Mr. Wells is a graduate of Iowa State College, where he received the B.S. degree in Electrical Engineering in 1942. Between 1942 and 1946 he was Associate Engineer at Wright Field Aircraft Radio Laboratory and was responsible for circuit development work on airborne IFF equipment. In 1946 he worked at Bendix Radio on the development of AM and FM communication receivers, and from 1946 through 1952 he was associated with the Glenn L. Martin Company as Project Engineer on airto-air missiles. He is an active member of the American Ordnance Association and the Association of the U.S. Army. •3•

(Continued on rage 90A)

WHEN WRITING TO ADVERTISERS PLEASE MENTION-----PROCEEDINGS OF THE IRE

March, 1960



A NEW COMPUTER TECHNOLOGY solving problems hitherto unsolvable

DYSTAC (Dynamic Storage Analog Computer) synthesizes the advantages of both analog and digital computers. Dystac combines the analog's speed, lower cost, ease of programming, and improved output data presentation with the digital computer's unique capacity for data storage and time sharing of computer elements. With Dystac you can now solve the most complex problems, ranging from distillation column design to multidimensional heat transfer... from boundary value problems to transport lags.

In distillation column design, for example, Dystac employs only 44 amplifiers on a four-component problem: the number of amplifiers required is independent of the number of plates. With any other analog computer each plate must be reproduced in the circuitry-for a 30-plate column, over 1200 amplifiers are needed unless manual reprogramming is employed. What's more, only

DYSTAC yields a plate-by-plate display of all column conditions.

In heat transfer studies, DYSTAC will speedily and accurately solve hitherto unsolvable partial differential equations. It also makes practical for the first time instantaneous solutions to such trial-and-error problems as automatic optimization, automatic correlations, data fitting, probability distribution. Fourier analysis, convolution and superposition integrals and eigenvalue.

When you buy a CSI computer with DYSTAC you are buying more than the finest general purpose analog computer. You buy additional capacity as well as capability. For when DYSTAC elements are not in use as memory devices they are available to the computer as operational amplifiers. Planning to interconnect digital and analog computers? DYSTAC is the answer.

See Dystac in operation at our plant. If you can't attend a demonstration write, phone or wire for full information.



PROCEEDINGS OF THE IRE



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ELECTRONIC SUB-SYSTEMS IN MEGPOTS IN GASEALS TOGGLE SWITCHES IN HERMETIC SEALING SERVICES



(Continued from page 88.4)

Professor Arthur W. Melloh (A'33– SM'45) of the University of New Mexico faculty has been named professor of electrical engineering and dean of The State University of Iowa College of Engineering by the State Board of Regents.

The appointment will be effective July 1. As head of the SUI engineering college he succeeds Dean Francis M. Dawson, who retired in 1957 after 21 years as dean. At the request of University officials Dean Dawson had continued serving while a successor was sought, but illness forced him to widhdraw from active duty in the college early last year.

The new Iowa engineering dean, 52, is a native of Minnesota, and earned the B.E.E., M.S. and Ph.D. degrees from the University of Minnesota, where he also served as instructor for four years.

His subsequent employment included two years as research engineer for Automatic Electric Co. of Chicago, three years with the San Diego Radio and Sound Laboratory, and two years as senior research engineer for Stromberg-Carlson Co. in Rochester, N. Y. From 1947 to 1956 he was vice-director of the Texas Engineering Experiment Station, a branch of Texas A & M College, and since 1956 he has been professor of electrical engineering at the University of New Mexico, Albuquerque,

(Continued on page 92.4)



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150-175 Mc Model H+150-2A

SAVE MONEY—USE ONE ANTENNA FOR SEVERAL CHANNELS

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PROCEEDINGS OF THE IRE

March, 1960

World Radio History

91A



BOOTH 2827—Radio Engineering Show—N.Y. Coliseum—Mar. 21-24

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(Telephone Type Lever or Key Switch)

- T-BEAM FRAME-provides rugged assembly.
- SINGLE HOLE MOUNTING—in panels up to ¼" thick.
- STANDARDIZED NON-TURN DEVICE.
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See Catalog No. S-58 or write for special Catalog S-302.



IRE People

(Continued from page 90.4)

Dean Melloh is the author of four instructional books for the U. S. Navy, three Experiment Station Bulletins, and five publications in technical journals. His professional memberships include the American Institute of Electrical Engineers, American Society for Engineering Education, Sigma Xi, Tau Beta Pi, and Phi Kappa Phi. He is a Registered Professional Engineer in the State of Texas.



Association Activities

The auditorium at Electronics Park Syracuse, N. Y., has been designated W. R. G. Baker Hall in honor of Dr. W. R. G. Baker, retired Vice President of General Electric, former Director of the EIA Engineering Department, and past President of the Association. One of the leading pioneers in the field of electronics. Dr. Baker now is Vice President for Research at Syracuse University and President of the Syracuse University Research Corp. He is also treasurer of the IRE. The announcement of the new designation of the auditorium was made at a luncheon held at Electronics Park in honor of Dr. Baker's 67th birthday. . . . A special EIA subcommittee studying color picture concepts advanced by Dr. Edwin H. Land, head of the Polaroid Corp., has urged that "Dr. Land and others be encouraged" to continue their basic studies, but recommended that the FCC color TV signal be left unchanged "at this time." The subcommittee, formed early last year by the EIA committee on Broadcast Television Systems, stated that the FCC signal "permits continued investigation of Dr. Land's method without deterioration of pictures reproduced by receivers making full use of the information present in the signal." The FCC color signal, the subcommittee pointed out in a statement of its findings, 'carries simultaneously the information for three-color reproduction and for methods outlined by Dr. Land" and "does not result in a large increase in the cost of receivers." The subcommittee's statement on the work of Dr. Land was signed by the BTS; Committee Chairman, Charles J. Hirsch. . . ELA has gone on record with the FCC as supporting the proposed reallocation of the 460-461 mc from the Citizens Radio Service to the Industrial Radio Services.

In a filing of FCC Docket No. 11959 (FCC Second Notice of Proposed Rule Making), the Association stated: "We be-

(Continued on page 94.4)

* The data on which these NOTES are based were selected by permission from *Weekly Report*, issues of December 21, 1950 and January 4, 11 and 18, 1960, published by the Electronic Industries Association, whose helpfulness is gratefully acknowledged.



Waters new pots conquer space

Two new 1/2'' Waters pots conquer a space problem for many a harassed space age engineer. Both require up to 25% less space behind the panel than pots having identical specifications. Available with terminals (shown), wire leads or printed circuit pins. Case lengths are only 3/8''. The new APS 1/2 is designed for bushing-type mounting. The WPS 1/2, designed for servo mounting, is the smallest potentiometer available for general use in rugged servo applications. Both are capable of dissipating 2 watts continuously! Reliability test reports available. Write for Bulletin APS-160.



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For Demonstrating and Testing Auto Radios TRANSISTOR OR VIBRATOR OPERATED 6 Volt or 12 Volt! New Models ... Designed for testing D.C. (mouels . . . Designed for resting D.C. Electrical Apparatus on Regular A.C. Lines. Electrical Apparatus un Rebuier Aux en Equipped with Full-Wave Dry Disc Type — Recliner, Assuring Noiseless, Interference-Free Operation and Extreme Long Life and Reliability. O.C. OUTPUT INPUT AMPERES SHIP USER TYPE A.C. Volts 60 Cycles VOLTS Cont. Int. 10 20 610C-ELIF 22 \$49.95 110 6 12 12 6 20 40 620C-ELIT 110 33 \$66.95 -01 112 10 20 SEE YOUR JOBBER OR WRITE FACTORY NEW MODELS \sqrt{NEW} DESIGNS \sqrt{NEW} LITERATURE "A" Baltery Eliminators + DC-AC Inverters + Auto Radio Vibrators

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(Continued from page 92.4)

lieve that the proposed reallocation of frequencies . . . represents forward-thinking in spectrum utilization and therefore urge its adoption."

Also recommended was transfer of 465.475-466.475 mc to the Industrial Radio Services as a companion megacycle with the Commission's proposed transfer of the 460–461 mc band.

Electronics Abroad

The Japanese government has decided to adopt the NTSC color TV system and formal announcement will be followed quickly by issuance of a station license to the NTV network. This will make Japan the second nation in the world, following the U. S., to have commercial color TV. Because the price of a color TV receiver will be almost a year's gross wages for the average factory or office worker in Japan even after mass production of fairly small sets is achieved, the chances are remote of color TV reaching even a fraction of the homes in Japan, the report from there stated. At present, it is estimated, one million black-and-white TV receivers have found their way into one Japanese home in 20. . . . Imports of electronic products into the U.S. during the first nine months of 1959 totaled in excess of \$48.8 million

and were more than two and one-half times those of the same period last year while other exports declined slightly, the Electronics Division, BDSA, reported. Among the increased imports were: Radio apparatus and parts. The rapid rise in these imports-from \$3.4 million in 1955 to \$28.2 million in 1958 took \$43.3 million during the first nine months of 1959- attributable principally to the increased shipments of radio receivers from Japan. Significant among current Japanese shipments of electronic products to the U.S. are radio receivers-principally transistor portable. Japanese shipments of radio receivers of all types (not included radio phonographs) to the U. S. numbered 641,208 in 1957; 2,506,920 in 1958 and 3,900,222 in first nine months in 1959; valued at \$5.3 million; \$17.9 million and \$37.5 million, respectively. Exports of record players and record changers to the U. S. from the United Kingdom alone were valued at \$7 million in 1957, \$9.2 million in 1958, and \$4.5 million in the first half of 1959. The U. S. is by far the most important single foreign market for UK electronic producers, accounting for \$17.2 million in 1958 compared with exports of \$8 million to Australia and \$7.8 million to Canada, the next largest market. UK imports of electronic products from the U.S. in 1958 were valued at \$8.1 million.

Engineering

The National Bureau of Standards has announced that the handbook PREFER-RED CIRCUITS, Navy Aeronautical Elec-

(Continued on page 96.4)



finished pieces per hour. Can be operated by unskilled labor. Easily set up and adjusted to different lengths of wire and stripping. ENGINEERING consultation without obligation. Machines for all types of wire lead finishing. VISIT US AT BOOTH 4205—

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Highly Uniform Rugged Rotary Switches, TROLEX Series. Exceptionally high uniform reliability is achieved by an entirely new manufacturing concept. For military and







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(Continued from page 94.4)

tronic Equipment, NAVAER 16-1-519, Supplement No. 2, 54 pages, dated April 1, 1959, is available from the Government Printing Office, Washington 25, D. C., price 30 cents, in looseleaf format. The supplement contains 5 circuits and a note on preferred regulated voltages for transistorized equipment. The aim of the preferred circuits program is the voluntary reduction of unnecessary circuit variations in military equipment. . . . Progress by one Navy engineering contractor in improving the thermistor infrared detector during five years of research and development is reviewed in a report just released to industry through the Office of Technical Services, U. S. Department of Commerce. Also available is a Navy report on problems and advances, to mid-1954, in development of photoconductive detectors for infrared atomic spectroscopy. The two volumes are: Thermistor Infrared Detectors: Part 1-Properties and Developments. (Order PB151767 from OTS, U.S. Department of Commerce, Washington 25, D. C., \$2.75.) Infrared Atomic Spectros-copy, Based on Use of Photoconductive Detectors. (Order PB 151953 from OTS, U. S. Department of Commerce, Washington 25, D. C., \$1.25.)

MHATARY ELECTRONICS

The Pentagon's list of the top 100 defense contractors in fiscal year 1959 is changing in composition compared with former years, a close examination reveals. A heavy preponderance of the first 25 contractors are engaged in missile or electronic production. General Dynamics won top place with total awards of \$1.6 billion or 7.2 per cent of the total of defense dollars. This compares with 6.4 per cent awarded to this firm in FY 1959. It is understood that over 80 per cent of the company's defense contracts involved missile-electronicaircraft programs. Boeing Airplane Co. dropped to second place with net contracts totaling \$1.167 billion; North American moved back up into third place with contracts totaling \$1.018 billion and it is understood that cancellation of the F-108 and the stretchout of the B-70 bomber will affect its position in later reports. General Electric remained in fourth place with \$914 million in new awards and its share of DOD dollars increased from 3.6 per cent in FY 1958 to 4.1 per cent in FY 1959. Lockheed Aircraft was placed fifth with \$898.5 million or 4 per cent of total defense dollars, and Douglas Aircraft was in 6th place with \$676.4 million in new contracts. United Aircraft, which dropped out of the top 10 in FY 1958, moved from 14th to 7th place with new awards totaling \$538.2 million or 2.4 per cent of defense dollars. The Martin Co., in the top 10 for some time, was placed 8th with \$524 million; Hughes Aircraft was ninth with \$494 million, and AT & T was 10th with \$476.5

(Continued on page 98.4)

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HALLIBURTON MILITARY SERIES 120H-Seamless drawn, heat treated aluminum, Series 120H Cases ... available in all sizes ... provide watertight, airtight and shockproof, heavy duty reusable cases for the carrying, transit and storage of aerial cameras, electronic controls and devices, aerological equipment, navigation instruments and other military equipment. The performance characteristics of these cases comply with the environmental conditions of humidity, extremes of temperature. fungus, salt laden atmosphere. shock, tumbling and submersion, in accordance with the provisions of Military Specification MIL-C-4150E. Series 120H Cases also conform to the construction and performance requirements of Specifications MIL-T-945A, MIL-STD-108C, Class I, Type Watertight, as electronic instru-ment and/or combination cases, MIL-T-4734, MIL-T-4807, MIL-E-4970, and other applicable standards.

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PROCEEDINGS OF THE IRE March, 1960

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Los Angeles: Ash M. Wood Co. Phone: CUmberland 3-1201





(Continued from page 96.4)

million. The next 10 contractors comprised a mixture of missile-electronic-aircraft manufacturers and were listed as follows: McDonnell Aircraft, \$403.5 million (not including the MERCURY program); Sperry Rand, \$403.2 million; Raytheon Co., \$392.6 million; Chrysler Corp., \$323.2 million; Grumman Aircraft, \$300.2 million; Republic Aviation, \$280.5 million; IBM \$276.9 million; Bendix Aviation, \$271.3 million; Westinghouse Electric, \$238 million; and General Motors Corp., \$210.7 million. The DOD reported that in FY 1959, aircraft and missile contracts accounted for 64 per cent of the value of awards of \$500,000 or more to these companies, and electronics contracts accounted for an additional 11 per cent. Also, the DOD said, 59 of the 100 companies are engaged directly in aircraft and missile work, or in electronics and R&D work directly related to aircraft and missile pro-The DOD has made public grams what it calls the first comprehensive guide to available military technical resources in a single broad area of technology-in this case, the field of missile ground support equipment. The directory is designed to encourage a maximum exchange of technical information in the military departments and in industry, with the prime pur-

(Continued on page 100.1)

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- Direct substitution measurements by means of broad-band impulse calibrator, without charts, assure repeatability.
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The unique design of Model NF-105, with 4 plug-in tuning units, avoids costly repetition of circuitry and components common to all frequency ranges, at savings in size, weight and cost. Simple to operate, this instrument permits fast and accurate measurements of both broadband or CW signals. Send for our Catalog 604.

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(Continued from page 98.4)

pose of avoiding the duplication of effort, the DOD said. The 15-page publication lists almost 200 items of equipment associated with missile GSE, with the name, location, and telephone number of the military personnel conversant with the latest developments in each field. Rhw publication is the most immediate and end-prodnet of a study on missile GSE which has been in progress since March, 1958. The publication, "Technical Resources Directory," may be obtained from the Office of Technical Services, Department of Commerce, at 50 cents a copy, and should be ordered by number—PB 161103.

OUTLOOK FOR 1960

A record-breaking year is in prospect for the electronic industries in 1960, according to the Business and Defense Services Administration, Commerce Department, and the outlook "is excellent for all product lines." The combined output of these industries, the roport states, "is expected to reach the \$10 billion level for the first time in their comparatively short history." A summary of the BDSA projections in the consumer, military, industrial, tubes-semiconductor, and parts industries follows:

Consumer Products-Factory sales of consumer type electronic products, al-

ready at a record high of \$1.95 billion, are expected to reach a level of \$2.2 billion in 1960. Military Electronic Equipment-The production of electronic equipment for military services and for the Space Agency should continue upward in 1960, at the rate of better than 15 per cent. Industrial Electronic Equipment-A good year lies ahead for manufacturers of commercial and industrial electronic equipment. Indications are that the demand for these products will continue upward, with gains in 1960 of at least 10 per cent over-all. Electron Tubes-Total factory sales of electron tubes are expected to be up 5 to 6 per cent in 1960 from 1959 levels. Semiconductor Devices-The factory sales of semiconductor devices-transistors, diodes, and rectifiers-rose by 75 per cent. from \$210 million in 1958 to \$370 million in 1959. Equally vigorous growth is expected in 1960, although at a lower rateabout 30 per cent. Shipments should approach a total of \$470 million by the end of the year. Other Electronic Components-All types of electronic components should share in the rise in output forecast for 1960. Factory sales increased nearly 20 per cent in 1959, and should increase further in 1960, by about 12 per cent, bringing the total for the year near \$1.8 billion.

REVIEW OF 1959

Year-End Statement by David R. Hull, President, Electronic Industries Association: At the close of business on December 31, the electronics industry will have

(Continued on page 102.4)

NEW GEARLESS MULTIPLE TRANSFORMER WINDER with Instant Spiral/Rapid Traverse

7 TIME-SAVING FEA-TURES: 1) Unique wire carriage quickly shifts into coils' margin, adds extra turns for lead purposes. 2) Shorter set-up time. 3) No gear changing-pitch selector permits instant selection of turns per inch range. 4) Instantly adjustable winding width. 5) Winding cycle instantly reset by touching counter lever. 6) Instant spiral/ rapid traverse device. 7) Front loading of tension



permitted by compact machine design.

Model 405-AM multiple winds paper section power, audio, fluorescent ballast and similar transformer coil types up to 9" OD if round, 412" OD if rectangular, 6" long, using wire sizes 16-44. Maximum center to center distance for adjustable wire guide rollers 2312", distance between head and tailstock 33" max. & 21" min., winding range 19-454 turns per inch, winding speed up to 2000 RPM, output end of spindle 7%" flatted shaft. Furnished with motor, counter, brake, tailstock, tilting table paper feed & 20 tensions with new support bracket permitting vertical, horizontal & angular tension adjustment.

Reduce winding costs and time with Model 405-AM.

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Sub-Fractional Horsepower Motors

Frequency and Time Standards

PROCEEDINGS OF THE IRE March, 1960

Micropot Potentiometers

TUBE PROBLEM:

The Armed Forces needed a new version of the 6J4 reliable tube type which would provide a tube life of almost 1000 hours. Existing tubes of this type had an average life of only 250 hours. In addition, this new tube had to be produced under ultra-high quality control standards.

SONOTONE SOLVES IT:

By making improvements in the cathode alloy and setting up extremely tight controls in precision, manufacture and checking, Sonotone engineers produced a 6J4WA with a minimum life of 1000 hours...most running much longer.

RESULTS:

The Sonotone 6J4WA is one of three reliable tubes now being manufactured under U.S. Army Signal Corps RIQAP (Reduced Inspection **Quality Assurance** Program), monitored by the U.S. Army Signal Supply Agency. And the same rigid quality standards apply to Sonotone's entertainment type tubes as well.

Let Sonotone help solve *your* tube problems, too.



Engineering Notes

(Continued from page 100A)

achieved an unbroken succession of new all-time highs in total factory sales for every year of the decade of the 1950s. We believe the 10-year record, which twice in the period ran directly counter to national economic recessions, has few parallels in the recent history of American industry. On the basis of information now available. the industry's total 1959 business, at the factory level, may be estimated at \$9.2 billion. This is more than 31 times the total for the first year of the decade and represents an increase of nearly 16 per cent over the previous high of \$7.94 billion established in 1958. Every major segment of the industry reached new peaks in sales during 1959. In only two other years of the last ten-1956 and 1957-were records established on a comparable across-the-board basis. Manufacturers of consumer products made a substantial comeback from the 1958 recession with a 1959 sales total of \$2.05 billion, or \$450 million better than last year. The 1959 figure of \$1.1 billion for replacement parts, tubes, and semiconductors is \$240 million higher than 1958. Industrial and military products set new highs for the tenth year in a row with totals of \$1.55 and \$4.5 billion. These were \$170 and \$400 million, respectively, over 1958. The industry's record as a creator of jobs

(Continued on page 104.A)



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* Six models, covering 1 to 18 kmc. Broadest frequency range available. Electronic sweep of RF output, or extremely stable CW operation.

* Linear frequency sweep coverage over all or part of each band for rapid evaluation of reflection coefficient, gain, attenuation and other network transfer characteristics.

* Two adjustable frequency markers for convenient calibration of oscilloscopes or recorders. Markers save valuable test time by indicating either band limits or intermediate frequency values. An exclusive Alfred feature on all models.

 \star 0.5 microseconds rise and fall response to AM – equivalent to a 2 megacycle band pass. Another exclusive Alfred feature.

* Quick Look readout. See frequency range, markers and sweep time at a glance. No cumbersome calculating.

* Fast sweep for flicker-free oscilloscope presentation; slow sweep for mechanical recorder operation.



FREQUENCY CONTROL: Continuously adjustable with direct callbrated dial. Calibration accuracy, 1%. POWER OUTPUT (minimum): 10 mw. Continuously adjustable from zero to maximum. VSWR (maximum): 2:1.

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SELECTOR: Recurrent sweep, single sweep, CW, and external on panel switch.

CONTROL: Single sweep, triggered by panel-button, or external positive go-ing signal 20 volts or greater. SWEEP WIDTH: Continuously adjustable

TIME: 100 to .01 seconds. MONITOR OUTPUT: Positive linear sawtooth, 45 volts peak; Blanking out, 75 volts negative. EXTERNAL SWEEP: 200 volts gives full

sweep width.

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INTERNAL SQUARE WAVE: RF output alternately 0 and unmodulated CW value. Frequency 800 to 1200 cps. EXTERNAL: 30 volts maximum signal increases RF output from 0 to maximum.

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to test small signal and saturation gain of Model 503 Traveling Wave Tube Amplifier. Microwave Leveler, Alfred Model 704, holds power output from oscillator constant within ± 1 db.

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(Continued from page 102.4)

also has been outstanding. The total employed in electronic manufacturing is now estimated in excess of 760,000. This is more than double employment in the industry at the start of the decade. Every state in the union and the District of Columbia now have at least one electronic plant, although more than half the industry's manufacturing employment is in the six states of New York, Pennsylvania, New Jersey, Massachusetts, Illinois, and California. The chances now appear to be excellent that 1959 factory sales records will be surpassed in 1960 in all of the industry's principal product categories. On the basis of studies by the ELA Marketing Data Department, a total 1960 sales forecast of \$10.35 billion now seems to be justified. This would represent a 121 per cent increase over the record highs of 1959. The increasing importance of the industry in the nation's program of defense production was demonstrated by the substantial rise in military product sales during 1959 in the face of a general levelling off of total defense procurement. The electronic content of military hardware has risen steadily during the decade. It now stands at about 33 per cent of total defense procurement and can be expected to increase fairly substantially next year. The consumer segment of the industry, manufacturers of its oldest line of products, showed real vitality in snapping back from a \$100 million decline during the recession year of 1958 to set a new all-time high with a 28 per cent sales increase. Gains were achieved in each of the main consumer product categories, both in units sold and sales dollars, with radio manufacturers enjoying their best year since the post-war boom of the late 1940s. It is anticipated that growing public awareness of the availability of extremely high quality home-reproduced music will contribute significantly to an increase of more than 10 per cent in consumer business for 1960. The electronic replacement business, it is believed, can look forward to an improvement in 1960 sales comparable to that of this year. This will reflect increases in the quantity of original equipment in use rather than any change in quality of products which, in fact, are steadily being made more reliable. The immediate and long-term future for industrial electronics looks particularly bright. For one thing, we have just begun to scratch the surface of market potentials in that portion of the industrial field represented by electronic computation and data processing. Experience with computers for the control of processes in petroleum refining and the production of chemicals, cement, and electric power foreshadows widespread use of similar equipment, within relatively few years, as a route to higher industrial operating efficiencies. Moreover, I am confident that, within the next decade, we shall see computers employed by business management for the making and execution of decisions to an extent undreamed of today. It has been recognized for some time that (Continued on page 106.4)



PHILCO, one of the world's leaders, demanded reliability, design versatility and production ease in the contact selected for the very "heart" of its Input-Output Processor—the matrix panel. And that is why PHILCO selected ELCO's Series 5201, from our Series 5000 board-to-board printed circuit Varicon connectors! Furnished on disposable plastic strips,

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Plug-in module, shown mounted on matrix panel utilizing Elco Varicon contacts.



PROCEEDINGS OF THE IRE March, 1960



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analysis, nuclear magnetic resonance, etc. The tollowing characteristics and features are available only with the PG-650-C.

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(Continued from page 104.4)

electronics will have a vital and exciting part in man's efforts to conquer space. This role recently was expressed in terms of future business in an EIA Marketing Data Department study. It showed hardware for non-military space applications approaching \$100 million last year and increasing more than eightfold by the end of 1970. A major growth factor in space electronics will be the development of new and better satellites for communications and other purposes which are certain to be produced successfully despite recent rocket failures and other setbacks. It is now generally recognized that, in venturing into space, the electronics industry has had to meet standards of reliability well above those which proved to be acceptable in more conventional undertakings. Improvements in this direction are certain to be adopted by the industry generally with the result that, in a few years, better and more desirable products will be available throughout the entire electronics line. This cannot fail to be stimulating to the industry. Perhaps even more significant growth can be anticipated from a high level of research activity and product development. In view of this industry's favorable position with respect to progress in space and its record of scientific achievement, a \$20billion a year business in electronics by the end of 1970 does not seem unrealistic.





AERONAUTICAL AND NAVIGATIONAL ELECTRONICS

Oklahoma City-December 16, 1958

"Ferrite Load Isolators," J. Winchels, CAA.

"Radar Microwave Link," H. J. Burton CAA.

Oklahoma City-February 10

"Electrical Systems for Jet-Powered Transport Aircraft," C. L. Daniels, FAA Aeronautical Center.

Oklahoma City-February 18

"The Application of Dielectric Materials to Microwave Absorbers and Luneberg Lenses," E. F. Buckley, Emerson and Cuming, Inc.

Oklahoma City-April 21

"Pancake VOR-Latest Developments," E. Foster.

Oklahoma City-June 9

"FAA Long Range Radar," G. McKinnis.

Oklahoma City-- October 27

"GEEIA-Rapcon Pre-Engineering-TACAN," B. Williams, A. Ranscher, J. F. Maser, All of Tinker AFB.

ANTENNAS AND PROPAGATION

Boston-October 29

"Correlation Techniques Applied to Antenna Pattern Control," H. E. Band-J. E. Walsh, Pickard and Burns.

Boston--November 16

"Tetragon Antennas," M. L. Rosenthal, General Electronic Lab.

"VHF Signal Level Measurements Along A 2,000 Mile Path." A. S. Orange, Air Force Cambridge Res. Center.

Boston-December 15

"Antenna System For The Harvard Meteor Radar Project," A. K. Rodman, Radiation Engineering Lab.

"Radio Technique Employed In The Study of Meteors," W. H. McDonough, Harvard College Obs.

Los Angeles-September 10

"Analysis and Prediction of Radio Signal Interference Effects Due to Ionized Layer Around A Re-Entry Vehicle," W. C. Taylor, Lockheed Missile and Space Div.

Los Angeles-November 12

"Plasma Reactions on Electromagnetic Waves," Tutorial Presentation, R. S. Elliot, UCLA.

(Continued on page 108.4)

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(Continued from page 106.4)

Audio

Baltimore—December 14

"A New Method or Elimination of Turntable Rumble, Wow and Fhitter," P. Weathers, Weathers Ind.

Boston—November 12

"Frequency Fidelity in Audio Systems," C. I. Mahne, R.L.E., M.I.T.

Automatic Control

Milwaukee—October 20

"Optimizing the Transient Performance of A Pneumatic Temperature Control System," J. P. Metzger, Milwaukee School of Engineering.

BROADCASTING

Fla, West Coast-November 18

"Communications at a Strategic Air Command Base," W. E. Smith, U. S. Air Force, McDill Air Force Base.

Philadelphia—December 10

"A Report on Field Tests," E. M. Creamer, Jr., Phileo Corp.

Circuit Theory

Los Angeles-December 15

"Parametric Amplifiers," A. D. Berk, Hughes Aircraft Co.

"Inertial Navigation" R. S. Carlson, Autoretics Div., North American Aviation.

Philadelphia-October 20

"Introduction to Active Circuits," B. K. Kinariwala, Bell Telephone Labs.

Philadelphia-November 2

"Single and Multiple Port Active Devices," J. Hilibrand, RCA Labs

Philadelphia-November 9

"Impedance Properties of Active Networks," J. H. Mulligan, Jr., New York Univ.

Philadelphia—Nov. 16

"Non-Reciprocal Networks," H. J. Carlin, Institute of Brooklyn.

Philadelphia-November 30

"Theory of Parametric Amplifiers," K. K. N. Chang, RCA Labs.

Philadelphia-December 7

"Masers, Low-Noise Systems, and Possible Future Developments," R. H. Kingston, Lincoln Lab.

(Continued on page 110.4)



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Hermetic Ceramic Terminals, Magnetron Wells, Sapphire-to-Metal Seals





(Continued from page 108.4)

Communication Systems

Oklahoma City-November 19

"U. S. Air Force Sage, (Semi Automatic Ground Environment)," Maj. Holt and Maj. Peel, U. S. Air Force, and 1 White, Western Electric Co.

Electron Devices

Boston-November 24

"Electron Devices in Particle Accelerations," E. T. Westbrook, High Voltage Engineering.

Washington, D. C .- November 23

"The Practical Utilization of Microwave Energy for Microwave Propulsion," W. C. Brown and M. Theodore, Raytheon.

ELECTRONIC COMPUTERS

Akron-December 15

"Microwave Computers," F. Sterzer, RCA.

Baltimore—December 16

"Analog to Digital Conversion Techniques," K. Bacon, United Aircraft Corp.

Houston-November 25

"Some Maniacs I Have Known," N. Metropolis, Univ. of Chicago.

Houston-December 15

"Rice Institute Computer," M. Graham, Rice Inst

Los Angeles-September 30

"Computer Technology in Russia," W. H. Ware, The RAND Corp.

"Computer Technology in Russia," P. Armer, The RAND Corp.

Los Angeles-October 22

"Paran and its Applications," Y. Hata, Kanamatsu New York Inc.

Los Angeles--November 19

"The Honeywell 800 System," D. E. Robinson, Datamatic Div. of Minneapolis Honeywell.

Los Angeles—December 17

"The Kerr Effect in Ferromagnetic Research" D. Treves, Weizman Institute of Science,

"Vacuum Evaporated Information Processing Subsystems," K. D. Broadbeat, Hughes Research Labs.

San Francisco-November 24

"Computer Activities in Japan," E. Goto, Tokyo Univ.

(Continued on page 112.4)

March, 1960

110A

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE





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(Continued from page 110.4)

San Francisco-December 15

"Circuit Philosophy and Design for the new ILLIAC Computer," W. Poppelhaum, Univ. of Illinois.

ENGINEERING MANAGEMENT

Boston-December 3

"The Incident Process," P. Pigors, M.I.T.

San Francisco-December 8

"Problems in Research and Development Management," J. Church, Booz, Allen, and Hamilton.

ENGINEERING WRITING AND SPEECH

> Los Angeles-September 17-18 Third National Symposium,

Los Angeles-November 19

"Specification Writing," R. Norton, Hughes Aircraft Co.

INDUSTRIAL ELECTRONICS

Oniaha-Lincoln-November 20

Demonstration of Braille Multimeter, Sliderule, and Stylus Notes, La Von Peterson.

Demonstration of 9,000 RCA Panels, and Tour of Radio Engrg. Inst., Chief Instructor R. L. Hill, and L. Kozicki and L. Sedlak.

INFORMATION THEORY

Los Angeles-December 15

"Statistical Mechanics and Information Entropy," J. M. Richardson, Hughes Aircraft Co.

INSTRUMENTATION

Long Island-November 17

"Instrumentation in Oceanography," R. L. Erath, Grumman Aircraft Corp.

Los Angeles—December 2

"Radiation Instrumentation Techniques," L. Gardner, Litton Ind.

Washington-November 16

"Design of an RC Filter for use at Very Low Frequencies," W. S. Campbell, David Taylor Model Basin.

"A Telemetering Torque and Horse-power Meter," M. W. Wilson, David Taylor Model Basin.

Washington-December 21

"An R. F. Voltage Standard for Receiver Calibration in the Frequency Range of From 2 to 1000 Megacycles," G. V.

1124

Sorger and A. L. Hedrich, Weinschell Engineering Co., Inc.

MEDICAL ELECTRONICS

Los Angeles-December 10

"The Coding of Information in the Central Nervous System," W. R. Adey, U.C.L.A.

Montreal-November 25

"Engineering in Medicine," J. Hopps, National Research Council.

Washington-December 3

"Physiological Instrumentation for Space Flight," W. Greatbatch, Taber In-strument Corp.

MICROWAVE THEORY AND TECHNIQUES

Baltimore-December 9

"Useful Propagation of Microwaves Well Beyond the Horizon," T. J. Carroll, Bendix Radio Div.

Boston-October 21

"Microwave Properties of Thin Magnetic Films," P. Tannenwald, Lincoln Lab.

Boston-December 9

"Fundamental Properties of Para-metric Amplifiers," P. Johannessen, Sylvania Co.

Los Angeles-December 10

"Survey of Parametric Amplifier Research at SRL," Stanford Res. Inst.

Omaha-Lincoln-December 3

"Radar Scattering From Certain Peri-odic Discontinuities," E. D. Denman, Midwest Research Inst

San Francisco-December 2

"Microwave Generation Using Ferrites," J. Shaw, Stanford Univ.

Washington-December 8

"Microwave Directional Filters (striplines)," L. P. Tuttle, Jr., Melpar Inc.

MILITARY ELECTRONICS

Indianapolis-October 29

"Responsibilities in Electronics," G. R. Fraser, U. S. Naval Avionics Facility.

San Diego-December 16

"Flight Safety Criteria at the Atlantic Missile Range as Related to the Atlas V & D Missile System Configuration," S. L. Ackerman, Electronic Products of Convair Astronautics.

NUCLEAR SCIENCE

Oak Ridge-December 17

"Electron Energy Concepts in Electronics." J. L. Blankenship, Oak Ridge National Lab.

(Continued on page 114A)

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(Continued from page 112.4)

PRODUCTION TECHNIQUES

Boston-November 30

"A Comparison of U. S. and European Components and Techniques," L. Kahn, Aerovox Corp.

Reliability and Quality Control

Fort Worth-November 10

"Component Quality Assurance and Reliability Programs," A. W. Wortham, Texas Instruments, Inc.

Los Angeles-November 16

Panel—"Reliability Engineering Training," E. P. Coleman, U.C.L.A. Panel—"Reliability Engineering Train-

Panel—"Reliability Engineering Training," H. G. Romig, Hoffman Labs.—R & D. Panel—Summary by I. Doshay (moderator).

Space Electronics and Telemetry

Detroit—December 2 "New Advances in Telemetry," Dr. Epstein, Tele-Dynamics Inc.

(Continued on page 116.4)



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March, 1960

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(Continued from base 114A)

Los Angeles-October 20

"Aspects of the Digilock Telemetry System," D. W. Boensel, Space Electronics Corp.

Los Angeles-November 17

"Explorer VI Telemetry-Telebit," R. E. Gottfried, Space Tech. Labs.

VEHICULAR COMMUNICATIONS

Los Angeles-November 19

"Precision Measurements with Heterodyne Meters," R. Boniarz, Gertsch Co. Tape Recorded Report on Geneva Conf., Kittner and Watkins.

Los Angeles-December 17

"Transistorized Mobile Receivers," V. Stineman, General Electric Co.



AKRON

"A New Radio Interferometer Tracking System," Dr. C. H. Grace, Smith Electronics, Inc. 11/17/59.

"Methods of Continuous Radar Performance Monitoring," L. H. Fisher, Polytechnic Research & Development Co. 12/8/59.

ATLANTA

"Visual and Acoustical Research at Murray Hill," Dr. E. E. David, Jr., Bell Telephone Laboratories. 1/4/60.

BALTIMORE

"A New Approach to the Elimination of Turntable Rumble, Wow and Flutter." Paul Weathers, Weather Industries, 12/14,59.

"Molecular Engineering," Dr. Gene Strull, Westinghouse Electric Corp. 1/11/60.

RENELUX

"The Development of Color Television In the United States," Dr. G. H. Brown, RCA, 11/10 59.

BINGHAMTON

"Industry's Future in National Defense," George Metcalf, General Electric Co. Plant tour of the General Electric Johnson City facilities, 11/17/ 50

"Status of Electronic Microminiaturization," N. J. Doctor, National Bureau of Standards. 1/11/ 60

BUFFALO-NIAGARA

"Automatic Car Controls," R, S. Cataldo, General Motors. 9/16/59.

"Education for the Electronic Age," Dr. Ernst Weber, IRE President. 10/6/59.

"Threshold Sensitivity of Radar Display Devices," Dr. Carl Miller, Cornell Aeronautical Labs. 12/17/59.

(Continued on page 118.4)

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(Continued from page 116.4)

CEDAR RAPIDS

"New Developments in Color Vision," Dr. W. L. Hughes, Iowa State University, 11/11/59.

"Ten Years Before the Masthead," E. K. Gannett, IRE Editor; "Theory and Application of Tunnel Diodes," Drs. A. V. Pohm and R. Mattson, Iowa State University, 12/10–59.

CENTRAL FLORIDA

"Polaris Missile," Rear Adm. Rahorn, U. S. Naval Office in Washington, 10/15/59.

"Engineering Curriculum and Accreditation," Dr. J. Kueper, Brevard Engineering College, Tour of Melbourne Telephone Exchange and Microwave Relay System with comments by M. R. Hardt, Southern Bell Telephone Co. 11/27/59.

"Bio Effect and Need for Better Scientific Communication," Col. G. Knauf, Patrick AFB, Election of officers for 1960, 12/17/59.

CINCINNATI

"Twenty Vears with Radio-Controlled Model Planes," Dr. W. A. Good, Johns Hopkins University: "Mission Considerations for Nuclear Propulsion of Aircraft and Missiles." A. A. Hafer, General Electric Co, 12/15/59.

CLEVELAND

"Theories of Stereophonic Sound, Demonstration," B. B. Bauer, CBS Undoratories, 12/10/59, "Silicon Controlled Rectifiers and Applications," E. W. Hookway, G. E. Co. 12/17/59. COLUMBUS

"Recent Developments in Maser Amplifiers," Dr. W. S. C. Chang, Ohio State University, 1/12/60.

Elmira-Corning

"Microelectronic Computer Concepts Aimed at a Coming Revolution," Prof. E. W. Fletcher, MIT, 10/26/59.

"The Cleaning of Electronic Devices and Materials," F. J. Biondi, Bell Telephone Laboratories, 11/23/59.

"Cornell Radar Telescope," Dr. W. E. Gordon, Cornell University, 12–14/59.

Emporium

Social Meeting-Ladies Night, 12/11 59.

FLORIDA WEST COAST

"Some of the Problems of Space Navigation," C. W. Benfield, Minneapolis-Honeywell Regulator Co. 12/16/59.

Fort Huachuca

"Computer Hardware," R. L. Manuel, IBM, 12/21/59.

FORT WORTH

"Telephone Science and National Defense," Glenn Scott, Sonthwestern Bell Telephone Company, 12/15/59.

GAINESVILLE

"Electronic Focusing of High Density Electron Beams," Dr. A. D. Sutherland, Sperry Electronic Tube Division, 1/13-60.

HAMILTON

"CN Telegraphs and Microwave," C. Bridgeland, Canadian National Telegraphs. 12 15/59.

HOUSTON

"The Rice Institute Computer," Dr. Martin Graham, The Rice Institute, 12–15–59.

(Continued on page 120.4)





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(Continued from page 118.4)

HUNTSVILLE

"Airborne Bio-Medical Instrumentation," J. T. Powell, Redstone Arsenal, 12/14/59,

INDIANAPOLIS

"Precision Airborne Navigational Methods," J. F. Genna, U. S. Naval Avionics Facility, 1/15–60,

ITHACA

"The Van Allen Radiation Belts " Prof. T, Gold, Cornell University, 10/1 59.

"The Use of Weibull Probability Paper in Reliability Studies," John Kao, Cornell University, 11/5/59.

"Bandwidth Compression by Means of Vocoders," Frank Slaymaker, Stromberg-Carlson, 12/10/59.

Long Island

"Oceanography," Dr. J. B. Hersey, Woods Hole Oceanographic Institute, Movie "Washington At Work," 12/8/59.

L1.BBOCK

Tour of Southwestern Bell Telephone Company plant. 12/22/59.

MILWAUKEE

"Solid State Circuits," Harvey Cragon and Charles Phipps, Texas Instruments, 10 20/59,

"Crystals in Yesterday's Beauties and Today's Technology." Alan Holden, Bell Telephone Labs Inc. 11/4/59,

"An Introduction to Single Sideband," E. W. Pappentuss, Collins Radio Co. 12/15/59.

MONTREAL

"Transistor Logic Circuits," T. G. Rankin, Sperry Gyroscope Co. of Canada, 12/16/59.

NEW YORK

"The Simulation of Learning Processes," A. G. Schillinger, Polytechnic Institute of Brooklyn, 11/4/59,

NORTHERN ALBERTA

"Selective Signalling for Mobile Telephone Systems," R. Usher, Alberta Government Telephones, 12/7/59.

Phoenix

"The Moscow IGY Conference," Dr. 11, Richter, Convair Astronautics, 12/15/59.

PRINCETON

"The Practical Utilization of Power Transmission by Electromagnetic Means," Dr. W. C. Brown, Raytheon, 12/10–59.

OKLAHOMA CITY

Tour of the Sylvania Tube Plant, Shawnee, Oklahoma, 11/10/59,

"Dew Line," Dr. C. A. Dunn, Oklahoma State University, 12–8–59.

SAN DIEGO

"Interplanetary Travel," K. J. Bossart, Convair Astronautics, 12/2/59.

SCHENECTADY

"The Future of Engineering Education," Prof. H. W. Bibber, Union College and Prof. W. R. Beam, R. P. I. 12, 8, 59.

R. P. I. 12/8/59.
 "Electrically Steered Antennas Using Ferrite,"
 Harold Shnitkin, W. L. Maxson Corp. 1/12/60.

(Continued on page 124A)

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March, 1960



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(Continued or page 136.4)

PROCEEDINGS OF THE IRE March, 1960

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(Continued from page 136.1)

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Whom and What to See at the Radio Engineering Show

March 21-24, 1960

New York Coliseum

٩.,

These pages list the exhibitors at the Radio Engineering Show, with a brief description of what each exhibitor is showing, and a list of company personnel manning the booth. In each listing the booth number is given. Almost all booths have a 4-digit number. The first digit indicates the floor, the second digit indicates the aisle (aisle numbers increase from south to north). A few booths have one or two digit numbers, preceded by the letter "M". These booths are on the mezzanine at the back of the first floor. The show is divided into sections of related products, to help you in finding the products of your primary interest as easily as possible. These sections are:

First and Second floors-Components.

Third floor—Instruments and Complete Equipment. Communications Equipment and Systems, Computers, and instruments for test and measurements, microwave equipment. Fourth floor—Production. Machinery, tools, and raw materials; fabricators and services.

ACF Industries Incorporated ACF Electronics Division Riverdale, Md. & Paramus, NJ. Booth 1113

John H. Fournier, George B. Shaw, Robert Young, John A. Curtis, W. H. White, M. M. Millette

White, M. M. Millette Industrial Components, Microwave Components. The Avion Division and the Nuclear Products-Erco Division of ACF Industries Incorporated have been consolidated into the new ACF Electronics Division, Headquarters of the new division is in Riverdale, Maryland, with plants in Riverdale and Paranus, New Jersey; sales office in Culver City, California; and the Electro-Physics Laboratory in Bladensburg, Maryland.

A'G'A Division, Elastic Stop Nut Corp. ot America, Booth 2343 1027 Newark Ave.

Elizabeth 3, N.J.

H. Bostrom, S. Knapp, J. Newman, J. A. Long Agastat Time Delay Relay, Qualified to military specifications and aircraft requirements. Agastat munature relay unaffected by voltage variation, instantaneously recycling; time settings from 0.00 sec, to 120 seconds, Hermetically sealed or dust tight housings AC or DC contacts carry inductive load of 2 amps at 30 volts de and 3 amps at 100 volts ac.

AMP Incorporated, Booths 2529-2531 Eisenhower Blvd. Harrisburg, Pa.

B. Connet, A. Curtis, W. Haas, D. Hajjar, T. Harris, W. Hildebrand, O. Holmes, F. Howell, T. Kerr, J. Lyter, J. Milter, J. Pierce, J. Rausch, J. Simpson, E. Spooner, C. Stoup, J. Taylor, H. Wasiele

Patchcord Programmine Systems –universal and shielded systems and accessories. Pin Borrds– for matrix programmine, Double Throw Instrumentation Switches–(80-1500 pole). Multiple Aperture Devices (MAD)–features digital and analog memory planes. Maintainable Electronic Component Assemblies (AMP-MECA)–New modular assembly technique. AMPin-eert–Line of rack and panel connectors incorporating solderless pins and sockets. TERMASHIELD– Line of splices and iterutes designed for critical applications employing any shielded wire.

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Elizabethtown, Pa.

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Use this issue of *Proceedings of the IRE* to make your plans well before you get to the convention and show. Decide which technical sessions and social events you want to attend, and what exhibits at the show you will find of most interest. Advance planning will save you a great deal of time and effort, and will insure that you do not miss seeing or hearing about that one new product or technique which may be of vital importance to you in your work during the next year. A.P.M. Corporation 252 Hawthorne Ave. Yonkers 5, N.Y. Booth 1229 Solins, a Milton Morse, E. Otto

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A.R.F. Products, Inc., Booth 3938 7627 W. Lake St. River Forest, Ill.

▲ Arthur H. Maciszewski, ▲ John J. Pakan, M. Z. Massel, A. Przedpelski, J. Skolnick, D. J. Hamilton, Eugene D. Cahu

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(Continued on page 142A)

▲ Indicates 1RE member.

* Indicates new product.

MansonLaboratories.Stamford.Connecticut, designed six GL-7390's into this modulator whose power capability is 78 megawatts peak and 300 kilowatts average.



Below are shown the approximate envelope sizes and power outputs of two thyratrons now in use in high-power radar, as compared to the new General Electric tube.

Type 1257	Type 5948	New G-E Development (GL-7390)
8½"x20"	5"x16"	6"x11"
Avg. Power 33KW Peak Power 33MW	Avg. Power 12.5KW Peak Power 12.5MW	Avg. Power 66KW Peak Power 33MW

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The new General Electric GL-7390 hydrogen thyratron, which has the highest known power handling capability of any hydrogen thyratron now available. can be shipped immediately from stock. Designed for high-power radar pulse modulators. the GL-7390 features metal-ceramic construction for great mechanical ruggedness, smaller size for important space savings. and ability to switch extremely high average and peak power.

The external anode and grid construction allows direct convection cooling of the anode and grid. Reduced anode and grid temperatures during operation minimize the possibility of arc-back and/or grid emission. Ceramic-metal construction provides a rugged envelope which enables the GL-7390 to withstand shock and vibration conditions beyond the limits of glass designs. The anode and grid are in the form of solid metal cups solidly brazed to the ceramic body. This is a far stronger design than conventional glass seals and lead supports.

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March, 1960
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1/2". Sealed, sub-miniature type with onepiece metal case and bearing. Completely enclosed. Solid terminals, integrally cored with molded covers. Rotation: 320° electrical, 325° mechanical, 360° continuous.



SERIES C-078

 $7_{\rm M}^{\prime\prime}$. Weight only $\frac{1}{2}$ ounce. Independent linearity: $\pm 1\%$ of total resistance is standard. Linear or non-linear windings on flat card. Fully enclosed. Tolerance: $\pm 5\%$ standard, $\pm 1\%$ on order.



SERIES C-178

1%". Sine-cosine units with peak-to-peak accuracies to 0.25%. Independent brush contacts on common shaft, 90° apart. Ganged types available. Also 2" and 3" diameters.

Panel Instruments

1.00

11>

Ruggedized . . . round or square — miniature high precision units meet reduced size and weight requirements of aircraft and electronic applications



SERIES 100

1". Accuracy ±3% at full scale. Non-magnetic calibration. Scale length, 0.738". Background markings black or white, lance pointer, sealed solder lug terminals, aluminum housing. Watertight to meet MIL-M-3823 specs.



SERIES SC-031

¹/₂". Rugged, microminiature sealed unit. Includes external pivot D'Arsonval movement and high flux density Alnico magnet. Optional mounting, face plate and hex nut.



SERIES 131

142". Ruggedized to withstand shock, vibration or thermal extremes. Meets MIL-M-10304 specs. Positive watertight seal of meter and terminal studs.

Write for detailed literature on complete lines.

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March, 1960



NOW-48-56 Gauge Wire Coils built to YOUR specifications

Whatever your application-from hearing aids to missile sys-tems-Deluxe Coils' new fine wire plant can supply the miniature coils you need . . . built to your specifications for precision and accuracy.

Deluxe Coils' newest facility spans 15,000 sq. ft. It is air and sound conditioned and completely equipped to produce all types of miniature fine wire coils, 40-47 gauge, ultra fine wire coils, 48-56 gauge, and components.

Write for information on Deluxe Coils' fine wire production capabilities—and how they can be put to work for you, right away.

DELUXE COILS,



Whom and What to See at the Radio **Engineering Show**

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(Continued from page 142A)

Acoustica Associates, Inc., Booth 4120 Fairchild Court

Plainview, N.Y.

Kurt F. Vogt, A Stanley Jacke, Maurice Howell, Al Paley, Wm. Katsara, Artie Liers, A Robert L. Rod, Robert E. Rolnick, Samuel Markel, Wm. Abourezk, Jeanne A. LaTourette, Anita L. Langhauser

Heavy-duty ultrasonic cleaners, general purpo-e ultrasonic cleaners, immersible trausducers, *ul-trasonic cleaner accessories (baskets, pump and filter systems, covers, etc.), *ultrasonic cleaning chemicals, liquid level switches, continuous li-quid level switches, and ultrasonic R & D programs

> **Actioncraft Products** 2 Yennicock Ave. Port Washington, L.I., N.Y. Booth 4041

J. F. Murphy, J. M. Pelikan, H. Emory, W. J. Murphy, B. Oglethorpe

Insulation Sleeving to meet MIL-1-631, MIL-1-7444B, MIL-1-3190. Laminated split sleeve wire markers. Insulation sleeving, silicone rubber sleeving, marked and cut to specifications, (AIRTEX) Specification electronic lacing cords.

Acton Laboratories, Inc., Booths 3840-3842

533 Main St. Acton, Mass.

Leroy C. Bower, Jr., John Forrest, Lawrence Beloungie

Beloungie Ultra low frequency phase meters^{*}, citizen Band Transceivers^{*}, FM Monitors^{*}, #453 Transmission and Delay Measuring Sets^{*}, Pre-cision Shaft Couplings^{*}, Adjustable Slip Clutches^{*}, Miniature Gear Drives^{*}, Compact Transistorized Phase Meters, Phase Standards, Impedance Meters, Precision Drives, and Ampli-fiers will be displayed.

Advanced Vacuum Products, Booths 1310-1316 See: Indiana General Corp.

Ad-Yu Electronics Lab., Inc. 247 Terhune Ave. Passaie, N.J. Booth 3705 ▲ Paul Yu, ▲ Annibale Lupi, ▲ Roland St. Louis, Oscar Santos



Type 205B1 Phase Detector

*Precision Millimicrosecond Phase Detectors, 15 me to over 1000 me; RF Phase and Ratio Meters; direct reading Phase Meters, 1 cps to 500 kc; sensitive Phase Meter, 1° full scale, 0.005° accuracy; Phase Shifters; Delay Lines; continuously variable, step variable, tapped; Re-lays, 40 Microwatt sensitivity, meet MIL speci-fications. lays, 40 fications.

(Continued on page 116.4)

▲ Indicates 1RE member. 1 Indicates new product.

March, 1960

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World Radio History

INC.

COMPACT, RELIABLE, VERSATILE ... this is P&B's miniature MH relay

The MH is not a new relay.

As a matter of fact, we've been building and selling this series for seven or eight years. Its reliability and exceptional longevity have been proved in business machines, airborne computers and a host of other products.

Engineers like its fast action, its small size, its light weight. They like the wide selection of contact forms... up to 18 springs (9 per stack, DC) as well as the fact MH relays can be furnished to switch loads ranging from dry circuit to over 5 amps at 115 volts, 60 cycle resistive.

A multiple choice of terminations add to the MH's versatility. This relay, for example, can be adapted for printed circuits, furnished with taper tabs or a long list of other terminals. Get all the facts by calling your nearest P&B sales engineer today.





Breakdawn Valtage: 500 valts RMS between all elements. Ambient Temperatures: -45° C to +85° C.(-65° C ta +125° C

on special order.) Shack: 30g an special arder.

Vibratian: 10g from 55 to 500 cps.; .065" max. excursions from 10 to 55 cps. on special order.

Weight: 2½ ozs. max. (open relay) Terminals: Pierced solder lugs; special lugs far printed circuits, taper tab (AMP #78).

CONTACTS:

Arrangements: Up to 9 springs per stack.

The relays below are variations of the MH relay structure.



---- RELAYS

MA LATCHING [#] Electrical latch, mechanical reset, Small, versatile and offered with selection of contact arrangements.



Contacts rated 60 amp. 28 volts DC non-inductive. Will carry 150 amp. surge far a duration of 0.3 secands.



Material: ½ " silver standard: Palladium ar gald allay alsa available.

Laad: Dry circuits to 5 amps @ 115V AC res.

COILS :

spec.)

Resistance: 22,000 ahms max. Pawer: 100 mw per mavable min. to 4 watts at 25° C max.(200 mw min. to meet max.shock/vibratian

Duty: DC: Cantinuous. AC: Intermittent (Two pole relay max.) open. Sealed units supplied with full wave rectifier inside can.

Valtages: DC: Up to 110 valts. AC: Up to 230 volts 60 cycles.



MH SEAL-TEMP Features sealed coil to minimize contact contamination. Available as hermetically sealed relay only.

P&B STANDARD RELAYS ARE AVAILABLE AT YOUR LOCAL ELECTRONIC PARTS DISTRIBUTOR





MAGNETIC INSTRUMENTS CO., INC. THORNWOOD, N.Y. A Subsidiary of Pyrometer Company of America, Inc.

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The Dyna-Twin, a new headset of superior quality and performance, superior noise abatement characteristics. Excellent fidelity. Being lightweight and designed for maximum comfort it is ideal for all binaural and monaural applications. Dyna-Twin is engineered to withstand the rigorous environmental problems of temperature, humidity, vibration and shock. Choice of 4 mikes.

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TELEX, INC. Electro Mechanical - Acoustic Division Telex Park • St. Paul 1, Minnesota

Whom and What to See at the Radio **Engineering Show**

(Continued from page 144A)

Aeroprojects, Inc. 310 E. Rosedale Ave. West Chester, Pa. Booth 4238 W. C. Potthoff, D. D. Kirkpatrick, E. B. Webb, W. N. Rosenberg, E. D. Haigler, ▲ C. DePrisco, B. A. Valocchi Ultrasonic Metal Joining Equipment, Featuring a new '600 Watt Sonoweld Unit for joining a variety of similar & dissimilar metals.

Aerovox Corporation, Booths 2603-2607 740 Belleville Ave.

New Bedford, Mass.

New Bedford, Mass. Frank Marshall, Charles E. Krampf, Cyrus Stonehill, James Krampf, Ralph Parker, A Louis Kahn, A Dr. Antonio Rodriguez, Arthur Warner, A Harvey Pickett, Ruppert Jarboe, Roy Roskilly, Abraham Kalstein, Per Bogh Henrikssen, Thomas Cary, Owen Wood, G. Robert Tinay, Paul Goley, Guy Gardner, Charles Snow, Henry Taylor

CEROL—Rolled ceranic capacitors with high reliability and high capacitance. CERAFIL— ultra miniature ceramic capacitors . . . High Quality, Long-Life electrolytic capacitors; Epoxy clad Plate Assemblies; Aeroglaze carbon-de-posited resistors; POLYCAP capacitors in paper and electrolytic types.

> Aetna Electronics Corp. Readington Road

North Branch, N.J. Booth M-1 A Joseph F. McDonald, John Perkins, Frank Hunter, Robert McDonald, Henry Buser, Rob-ert We.:tworth



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Servo Analyzers .0008 to 100 CPS, Transmit-ter Inductors & Capacitors, Ultrasonie Cen-Ultrasonic Gen-Magnetostriction Inductors or Tuning Capacitors. Inductors, erator Tur Transducers.



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> Be sure to see all four floors!

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March, 1960

Simplified block diagram of Model CF-1. Amplitude and phase input functions are plotted on graph paper for presentation. Integration is observed on a dc oscilloscope. Absolute magnitude is recorded on any S-A Series 121 or APR 20 Antenna Pattern Recorder with a logarithmic response.



A sophisticated solution to the vexing problem of solving bounded Fourier integrals quickly and accurately, Scientific-Atlanta designed the Model CF-1 especially for the antenna design engineer.

The computer has broad general application including determination of the far fields of aperture antennas from the distribution of the field in the aperture, the far fields of arrays from the magnitude and phase of the currents in the elements, the frequency spectra of voltage pulses, and other physical problems involving Fourier transforms and their inverse transforms over finite limits.

PRICES

Model CF-1 Fourier Integral Computer...\$9,000

Model APR 22 Antenna Pattern Recorder (logarithmic response) ... \$4,300

See the CF-1 and other new S-A Microwave Instrumentation at IRE Booth 3909



Consult your nearby S-A engineering representative for more information. Or you may write directly to the factory for complete specifications. Address Dept. 433.

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PROCEEDINGS OF THE IRE A

March, 1960

superb new NULL DETECTOR



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Ranges: 11 linear ranges in 1x and 3x steps, from 100 μ v to 10 v f.s.; 5 non-linear ranges, 0.001 to 10 v f.s., each covering three decades.

Accuracy: Linear ranges, $\pm 3\%$ of f.s.; non-linear, $\pm 10\%$ of input.

Input Resistance: 10 megohms on all ranges. Max. power sensitivity over 10-17 watt.

Response Speed: On 100 µv range, 2.5-sec.; I-sec. on all others.

Noise: Below 2% f.s. all ranges.

Zero Drift: Less than $10 \mu v$ per day.

Output: 10 volts at 1 ma f.s.

Price: 151 Cabinet Model . \$395.00 151R Rack Model . \$385.00

Booth 3907, IRE Show



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INSTRUMENTS



Whom and What to

See at the Radio

(Continued on tage 152.A)



6

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-Leading in the application of novel printed circuit techniques to achieve a new order of miniaturization and compact design, Melpar is responsible for many innovations and new concepts in data transmission links, microwave receivers, communication terminal equipment, secure communications devices, and remote control systems. Melpar maintains extensive field operations, including a facility at Tucson, Arizona, where avionic equipment and systems are studied, installed, and tested for the Army's Electronic Proving Ground (Fort Huachuca).

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the need for a partnership between electronics and physical sciences, Melpar has established a materials research laboratory within its electronics complex. Research on the structure and application of new materials to support electronics is advancing at Melpar. Experienced research staffs are now evolving practicable, workable designs in such areas as high temperature effects on materials and molecular electronics.

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... proved ability, experience assure electronic achievement

MELPAR'S ENGINEERING DIVISION makes a continuing creative contribution to electronic development. And a number of significant electronic achievements have resulted from Melpar's imaginative approach to project and system engineering.

Electronic achievement at Melpar resulted in the conception, design, and production of one of the <u>largest</u> airborne electronic systems ever developed—the <u>first</u> integrated electromagnetic reconnaissance system . . . and <u>two</u> of the largest ground electronic systems—a ground based reconnaissance data handling system, and the F-101B weapon system simulator.

MELPAR EQUIPMENTS form an integral part of many advanced weapons systems, and equipment developed at Melpar will comprise a part of the first manned-satellite launched into orbital flight.



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SEND FOR CATALOG 359A-1 describing items above, also seals for circuit breakers, indicator lights and other electronic components.



Visit Our Booth 1229 at the IRE Show

152A

Whom and What to See at the Radio Engineering Show

(Continued from page 148A)

Airpax Electronics Incorporated Seminole Division Ft. Lauderdale, Fla. Cambridge Division Cambridge, Md. Booths 2306-2308 A.H. A. Cook, A.D. H. Holdt, J. W. Sullivan, A.W. Heister, W. Kouzoulas, A.D. A. Robinson, A.T. Dee, B. Linthicum, A.L. Gorder, M. Everhart, F. Marsh, J. Griffin, M. Rogers



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Electro-mechanical and transistor choppers^{*}, electro-magnetic circuit breakers, magnetic amplifiers, power, audio and pulse transformers, frequency detectors, tachometers, stretch indicators^{*}, demodulators, power amplifiers and meters. Hermetically sealed, miniature circuit breakers provide protection in critical circuits Short or long time delays available, operation from DC, 60 & 400 CPS.

Airtron, Inc., Booths 1610-1618, 1709-1717

See: Litton Industries, Inc.

Aladdin Electronics Division, Aladdin Industries, Inc., Booth 1918 703 Murfreesboro Road Nashville 10, Tenn.

F. G. Bassler, ▲ W. W. Stifler, ▲ P. E. Dicker, ▲ C. L. Freel

Ferrite Core Inductors. Miniature & Micro-Miniature Pulse Transformers, Wide-Band Transformers, High Quality I.F. Transformers, Missile I.F. Transformers.

Alden Electronic and Impulse Recording Equipment Co., Inc., Booth 1512 P. O. Box 125—Washington St.

Westboro, Mass. Milton Alden, A John Alden, Edward Cross, George Stafford, Lawrence Farrington Direct Graphic Recording with Alden "Flying Spot" Component Recorders for instant tone shade recordings in scanning, sampling or time base data collection. Precision Graphic Recorder for underwater sound; Ceilometer Recorder for underwater sound; Ceilometer Recorder for cloud height. Recorders for flaw detection. Spectrum Analysis, Radar, Infrared, Sonic, and Ultrasonic.

Alden Products Co. 117 N. Main St. Brockton 64, Mass. Booths 1508-1510 Russell Hawkins



PRODUCTS COMPANY

Building Block Components to Mount, House, Fasten, Connect and Monitor Electronic and Electrical Circuitry. Plug-In Chassis and Miniature Circuit Packages, Terminals and Cards, Cabinets, Miniature Indicating Lights, "IMF" Connectors, Detachable Line Cords.

(Continued on page 154A)

last year at the IRE Show

RIDER

introduced PRESSMAN'S TRANSISTORIZED CIRCUITS FOR DIGITAL COMPUTERS

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High Quality High Performance Extreme Reliability

From the leading manufacturer of power transistors, new Silicon Power Rectifiers to meet your most exacting requirements. Even under conditions of extreme temperatures, humidity and mechanical shock, these diffused junction rectifiers continue to function at maximum capacity! Thoroughly dependable, completely reliable—new Delco Rectifiers are an important addition to Delco Radio's high quality semiconductor line.

Conservatively rated at 40 and 22 amperes for continuous duty up to case temperatures of 150°C.

	TYPE	AVG. DC CURRENT	PIV	NORMAL MAX. TEMP.	MAX. Forward drop	MAX. Reverse current
	1N1191A 1N1192A 1N1193A 1N1194A 1N1183A 1N1184A 1N1185A 1N1185A	22A 22A 22A 22A 22A 40A 40A 40A	50V 100V 150V 200V 50V 100V 150V 200V	150°C 150°C 150°C 150°C 150°C 150°C 150°C 150°C 150°C	1.2V at 60 amps. 1.2V at 60 amps. 1.2V at 60 amps. 1.2V at 60 amps. 1.2V at 60 amps. 1.1V at 100 amps. 1.1V at 100 amps. 1.1V at 100 amps.	5.0 MA 5.0 MA 5.0 MA 5.0 MA 5.0 MA 5.0 MA 5.0 MA
14-20 W - 2A					inte at too ampor	at 150° C case temper- ature and rated PIV

For full information and applications assistance, contact your Delco Radio representative.

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Division of General Motors · Kokomo, Indiana





Whom and What to See at the Radio Engineering Show

(Continued from page 152A)

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▲ Fred W. Kruse, Jr., ▲ Paul N. Fulton, ▲ Gene Brakeman



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Alite Div., Booths 2205-2207 See: U.S. Stoneware Co.

All Products Co., Booth 2835 P. O. Box 110 Mineral Wells, Texas

▲ John Dunlavy, ▲ Tom Smith, ▲ Jim Buzbee, Carl Kessler

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(Continued on page 156A)

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I or **6**... How do you take the temperature of an electronic system? Use ONE Compact, Accurate, Lightweight, Completely Transistorized Oster-Developed Temperature Indicating System. Replaces 6 Indicating Systems.

100

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The device illustrated senses and displays the highest of 4 temperature sensor signals on the Convair B-58

supersonic bomber. It also reads out the temperature of the plane's 2 refrigeration systems. However, the unit can monitor temperature of any military or commercial electronic system. Special versions can be furnished to monitor any analog voltage signal.



Entire unit (component parts illustrated) has an accuracy of $1\frac{1}{2}\%$, weighs only 3.85 lbs. max., requires only $2\frac{3}{8}$ " panel space, and does the work of 6 indicating systems. Component parts consist of 4 sensors, a computer package with 6 channels, and a hermetically sealed indicator package.

The unit is typical of Oster designed equipment.







Computer Package

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For help with your instrumentation and display problems, talk to the specialists at the Avionic Division.

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Write for Data Book with specifications on Gudelace and Gudebrod's complete line of braided lacing tapes and dial cords-Temp-Lace, Stur-D-Lace, and Gude-Glass.

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GUDELACE Whom and What to See at the **Radio Engineering Show**

(Continued from page 15421)

Allegheny Electronic Chemical Co., Booth 4023 207 Hooker-Fulton Building Bradford, Pennsylvania Don Atkinson, L. J. Droege, N. J. Egli, R. L. Leslie

Silicon in all forms—Densified Chunk. Billets, Cast Rods, Seeds, Doping Alloys, Single Cry-stals, Slices and Special Forms.

Allegheny Ludlum Steel Corp., Booths 2509-2515

See: The Arnold Engineering Co.

Allen-Bradley Co. 136 West Greenfield Ave. Milwankee 4, Wis-Booths 2314-2316

▲ Wm. W. Garstang, ▲ George W. Vateı, A. Pfister, C. Dickinson, H. Za-bei, ▲ B. Teilkamp, ▲ H. Schlicke, ▲ E. Schwartz, R. Hower, ▲ B. Budny, P. Leow Subminiature components; rectilinear ad-

Subminiature components; rectifinear ad-justable fixed resistors; narrow band pass filter for UHF telemetry frequen-cies; ferrites; ceramic capacitors; ceramic filters; fixed composition and metal film precision resistors.

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Alpha Metals, Inc., Booth 4328 56 Water St.

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American Bosch Arma Corp., Booths 3910-3914

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(Continued on page 159.4)

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(Continued from page 156A)



American Electronic Laboratories, Inc. 121 N. Seventh St. Philadelphia 6. Pa. Booth 3935 Robert D. Freedman, Hamilton Priday, George Landfear, Maurice Polayes, Tom Burke



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American Lava Corporation, Booth 4502 Cherokee Blvd. & Mfgrs. Road Chattanooga 5, Tenn.

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(Continued on page 160A)

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INDUSTRIAL ELECTRONICS AND CONTROL, Second Edition

By R. G. KLOEFFLER, Kansas State Univ. Spans the whole field with a complete survey of electronic theory. Initial approach is through solid state theory rather than theory of tubes. Gives material on magnetic amplifiers, computer principles. instrumentation, etc. 1960. 540 pages. \$10.00

HYDROMAGNETIC CHANNEL FLOWS

By L. P. HARRIS, GE Research Laboratory. Analyses for three flows of viscous. incompressible, electrically conducting fluids in high-aspect ratio rectangular channels, subjected to transverse magnetic fields. A Technology Press Research Monograph, M.I.T. 1960. 96 pages. \$2.75

SPACE TECHNOLOGY

Edited by H. S. SFIFERT, Space Technology Laboratories, Inc. First to cover all phases seriously-ballistics and flight dynamics; propulsion; communications; man in space; scientific uses of space. 1959. 1188 pages. \$22.50

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Edited by E. M. GRABBE, S. RAMO, and D. E. WOOLDRIDGE, Thompson Ramo *Wooldridge Inc.* Full details on design of analog and digital computers and applications in science, engineering, and business. 1959. 1093 pages. \$17.50. Vol. 1, Control Fundamentals. 1958. 1020 pages. \$17.00. Vol. 3, Systems and Components. In Press.

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(Continued from page 159A)

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American Machine & Foundry Co., Potter & Brumfield Div., Booths 2702-2704 See: Potter & Brumfield, Inc.

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Whom and What to See at the Radio Engineering Show

(Continued from page 160A)

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Amersil Quartz Division, Booths 2110-2118

See: Engelhard Industries, Inc.

(Continued on page 164A)

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(Continued from page 162A)

Amperex Electronic Corp. 230 Duffy Ave. Hicksville, L.I., N.Y. Booths 2522-2524

F. Randall, E. Dorgelo, ▲ S. Gertzis, ▲ R. La Plante, J. Messerschmitt, ▲ I. Rudich, E. Bailley, B. Kutny, E. Feinberg, C. Roddy, W. Sandberg, A. Peterson



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A. J. Schmitt, J. F. Leach, R. W. Felber, R. F. Dorrell, J. Aylward, W. Jones, W. H. Rous, R. J. Gallagher, C. C. Camillo, C. F. Mikl, W. Adams, R. Klenert, M. L. Devine, R. M. Soria, R. E. Hall, H. Motz, H. P. Brontsema, R. Meade, R. Cobbin



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(Continued from page 164.4)

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(Continued on page 168A)

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 166A)

Ardente Acoustic Laboratories, Ltd., Booth 1820 See: British Radio Electronics Ltd.

Armed Forces Communications & Electronics Association. Booth 4226A See: SIGNAL Magazine.

Arnold **Engineering Co.** P.O. Box G Marengo, III. Booths 2509-2515

▲ Robert M. Arnold, C. S. Brand, A. C. Brown, R. Carroll, F. Dougherty, ▲ B. Falk, J. Kavanagh, B. Kramer, ▲ H. A. Lewis, J. E. Mitch, J. L. Jones Lewis, J. E. Mitch, J. L. Jones Permauent magnets of Alnico and Ar-nox, Silectron transformer cores, high permeability tape wound cores of Delta-max, Permalloy, Supermalloy, Super-mendur, as well as bobbin cores, Powder cores of Molyhdenum Permalloy, car-bonyl iron and Sendust; Barium titanate transducers, special magnetic materials

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(Continued on page 172A)

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NOTABLE ACHIEVEMENTS AT JPL ...



From MICROLOCK to microlock

One of the most interesting and useful scientific activities at JPL has been the development of MICROLOCK, a radio tracking and communication system for satellites.

Microlock is designed to transmit information over extreme ranges of space with a minimal amount of transmitter power and weight. The objective was achieved by sophisticated design of the ground receiving equipment. The design utilizes basic electronic circuits and techniques carefully combined in a novel manner to provide superior performance and sensitivity.

The satellite transmitter consists of a radio-frequency oscillator, phasemodulated by telemetering signals, and radiates a power of 3 mW. It is capable of operating for several months on a battery weighing one pound.

Used successfully in previous space vehicles, microlock remains a useful and expandable instrument for continuing space exploration. It is a prime example of JPL's activity on the space frontier.



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98.5% survival

That's the life-test record of General Electric low-current

GERMANIUM Low Current		FEATURES	JEDEC or G-E TYPE NO.	PIV.	MAXIMUM Ide ot T°C	MAXIMUM I Cycle (60 cps) Surge	MAXIMUM Storage TEMP. °C
			1N91	100	150mg at 55 gamb.	25A	85 *
		Alloyed junction type combining very	1N92	200	100mg at 55° amb.	25A	850
	Å	low forward resistonce with high bock	1N93	300	75mg at 55° amb.	25A	85°
	~	resistance	115N 1N93	300	75mg at 55° amb	254	850
	11		1N151	100	S00mg at 55° amb.	25A	85°
	1.00		1N152	200	500mg at 55° amb.	25A	850
		Single and double-fin Units	1N153	300	\$00ma at 55° amb.	25A	85°
			1N158	380	500 mp at 55° amb.	25A	85°
			1N315	100	100ma at 85° amb.	5A	95°
		Designed for high operating	U5AFIN315	100	100ms at 85° amb	5.4	100.0
		temperatures and low reverse current	1N368	200	100-c + 960-b	104	850
			11000	200	rooma er eg ame.	IVA	0.5
			IN 536	50	500ma at 100 amb.	ISA	175°
SILICON			IN537	100	500ma at 100 amb.	15A	1750
LOW			1N538	200	500ma at 100 amb.	15A	1750
CURRENT		Designed for maximum farward	USAFIN538	200	500ma at 100°amb.	15A	175
		conductance at high operating	IN539	300	500ma at 100 amb.	154	175
	1	temperatures (165°C)	IN540	400	S00me at 100 amb.	IDA ICA	175
			USAFIN540	400	500ma at 100*amb.	ACT	1/3
			1N 547	600	500ma of 100 amb.	15A	175
	1		IN1095	500	425ma at 100 °amb.	15A	1750
			1 N 10 96	600	350ma at 100 amb.	15A	175°
			1N440	100	300ma at 100° amb.	15A	175
			1N440B	100	500ma at 100 °amb.	15A	1750
	1		1N441	200	300ma at 100 amb.	15A	1750
	1		1N441B	200	500ma at 100 amb	15A	1750
	1 1	Similar to 1N536 series but with very	1N442 ·	300	300ma at 100°amb.	15A	175
		low reverse current. Ideal for magnetic	1N442B	300	500me at 100 amb.	15A	1/5-
		amplifier applications.	IN443	400	300me at 100°amb.	ISA	1/5*
			1N443B	400	500ma at 100 amb.	ISA	1/5
			1N444	500	300ma at 100°amb.	15A	175
	- i		1N444B	500	425ma at 100 amb.	15A	175°
	1.		1N445	600	300ma at 100°amb.	15A	175°
	e .		1N445B	600	350ma at 100 amb.	15A	175°
			1N1487	100	370mo at 100 amb.	15A	150*
			1N1488	200	370ma at 100 amb.	15A	150°
	2	Less expensive versions of 1N536 series for lower temperatures (140°C)	1N1489	300	370ma at 100 amb.	15A	150°
			1N1490	400	370ma at 100° amb.	ISA	150
			1N 1491	500	300ma at 100°amb.	15A	150°
			1N1492	600	250mo of 95° amb.	15A	150°
	· ·	Lower current and temperature	1N1692	100	600ma at 100°C amb.	20A	1250
		operation (100°C) than any of above	1N1693	200	600ma at 100 °C amb.	20 A	1250
		sorios very economical	1N1694	300	600ma at 100 °C amb.	20A	1250
		series; very economical	1N1695	400	600mo at 100 °C amb.	20 A	1250
			1N1100	100	500ma at 100°amb.	15A	1750
		Similar to 1N440B series	IN1101	200	500ma at 100°amb.	15A	1750
		Similar to treased series	IN1102	300	500ma at 100°amb.	15A	1750
			1N1103	400	500ma at 100 amb.	15A	1750
			1N599(A)	50	400ma at 100 amb.	10A	1750
		INISOD as sheet sheeth and INIS 40 as at	1N600(A)	100	400ma at 100°amb.	10A	1/50
		INDYY series similar to IND4U series;	1N601(A)	150	400ma at 100 °amb.	10A	1750
		INDYYA series similar to IN4408 series.	1N602(A)	200	400ma at 100 amb.	AOI	1/5
		Forward current ratings are somewhat	1N603(A)	300	400ma at 100 "amb.	10.4	1/5-
		lower.	1N004(A)	400	400ma at 100 amb.	104	1/3
			1N(005(A)	000	euuma at IVU amb.	104	1/3
			INDUG(A)	000	400ma et 100-amb.	IUA	1/5
			IN1115	100	1.5 A at 85° stud.	15A	175 2
SILICON		Same as 1N540 series except stud	IN1116	200	1.5A at 85° stud.	15A	1750
LOW		mounted; maximum forward conductance	1N1117	300	1.5A at 85°stud.	15A	1750
CURRENT		at high operating temperatures	1N1118	400	1.5A at 85° stud.	15A	175°
	0	at mgr apt and a trapped	1N1119	500	1.5A at 85°stud.	15A	1750
			1N1120	600	1.5A at 85° stud.	15A	1750
	4		1N253	95	1000ma at 135°stud.	4A	150°
	C.R	One of the first stud series,	1N254	190	400ma at 135°stud.	1.5A	150°
		JAN1N256 units available	1N255	380	400ma at 135°stud.	1.5A	150°
	T		1N256	570	200ma at 135°stud.	1.5A	150°
	13		1N550	100	800ma at 135°stud.	15A	175°
	w.	Come on 1NI4/OR sector among stud	1N551	200	800ma at 135°stud.	15A	175°
		Same as IN4400 series, except stud	1N552	300	800ma at 135°stud.	15A	175°
		mounted; extremely low reverse current;	1N553	400	800ma at 135°stud.	15A	175°
		well suited for magnetic amplifiers	1N554	500	600ma at 135°stud.	15A	175°
			1N555	600	600ma at 135° stud.	15A	175°

at 25,000 hours!

germanium rectifiers (Type 1N92); and silicon (Type 1N538) is even higher for 10,000 hours

General Electric low-current rectifiers have earned a reputation for reliability without equal in the industry. The table below is just a sample of the numerous life test studies which prove out the superior reliability *built into* all G-E rectifiers.

Maximum Forward Conductance

General Electric low-current silicon and germanium rectifiers are designed for maximum forward conductance at high operating temperatures. High current loads are carried *without* external heat sinks. Reverse current at maximum junction temperature is maintained at an extremely low level, making these devices ideal for low-leakage applications.

Minimum Forward Voltage Drop

Minimum forward voltage drop and a hermetically sealed case have combined to produce low-current rectifiers whose reliability exceeds all known existing MIL specs. A comparative study shows that these G-E devices have the highest resistance to thermal runaway at maximum full-load operating temperatures of those products tested. Choose the performance range you require from one of the most comprehensive low-current rectifier lines in the industry (see chart at left). Complete specifications are available from your General Electric Distributor or G-E Semiconductor District Sales Office. In Canada: Canadian General Electric Co., 189 Dufferin St., Toronto, Ontario. Export: International General Electric Co., 150 E. 42nd Street, New York, N. Y.

General Electric rectifiers are in stock at your local G-E Distributor

Survival Data From Operating and Elevated Storage Tests								
Type of Unit	PIV	Current (ma)	Type of Test	No. of Units	*Percent Survival			
1N92 Germanium	200∨	100	Operating at full load	69	98.5 @ 25,000 hrs.			
1N538 Silicon 200V 250 Operating 83 99@ at full load 10,000 hr plus elevated storage life								
*Percent survival == no. of good units x 100 total no. tested								

SILICON LOW CURPENT	FEATURES	JEDEC or G-E TYPE NO.	PIV.	MAXIMUM Idc ot T°C	MAXIMUM 1 Cycle (60 cps) Surge	MAXIMUM Storage TEMP. °C
CORRENT		IN 332	400	400ma at 150°stud.	15A	170 °
<u></u>		IN333	400	200ma at 150°stud.	10A	170 °
6		IN 334	300	400ma at 150 ^a stud.	15A	170°
GE		IN335	300	200ma at 150°stud.	10A	170°
		IN336	200	400ma at 150 ^a stud.	15A	170 °
		IN337	200	200mo at 150 ^a stud.	10A	170 °
		IN 339	100	400ma at 150°stud.	15A	170°
(E)	A widely used line similar in most	IN340	100	200ma at 150°stud.	10A	170 °
-	respects to 1N1115 series	1N341	400	400ma at 150° stud.	15A	170°
		IN342	400	200ma at 150°stud.	10A	170°
		IN 343	300	400ma at 150°stud.	15A	1,70°
		IN344	300	200ma at 150°stud.	10A	170 °
		IN345	200	400mo at 150°stud.	15A	170°
		IN346	200	200ma at 150 ^o stud.	10A	170 °
		IN348	100	400ma at 150°stud.	15A	170°
		1N349	100	200ma at 150°stud.	10A	170°
		IN607(A)	50	800ma at 135°stud.	15A	170 °
		IN608(A)	100	800ma at 135°stud.	15A	170 °
		IN609(A)	150	800ma at 135°stud.	15A	170 °
	IN607 series similar to 1N115 series;	IN610(A)	200	800ma at 135°stud.	15A	170.0
	1N607A series similar to 1N550 series	IN611(A)	300	800ma at 135°stud.	15A	170 0
		IN612(A)	400	800me at 135°stud.	15A	170°
		IN613(A)	500	600ma at 135°stud.	15A	170.0
		IN614(A)	600	600ma at 135° stud.	15A	170 *
		1N2 154	50	25A at 145°stud.	300 A	200°
	Stud Mounted Cells, Designed for 2 to	1N2155	100	25A at 145°stud.	300A	200 °
SILICON	20 ampere range High junction	1N2156	200	25A at 145°stud.	300A	200.°
CURRENT	temperature ratings, ungri forward	1N2157	300	25A at 145°stud.	300A	200 °
	velage date and the well estimate	1N2158	400	25A at 145°stud.	300 A	200 °
	voirage arop and thermal resistance.	1N2159	500	25A at 145° stud.	300 A	200 °
		1N2160	600	25A of 145° stud.	300.A	200.0



Semiconductor Products Dept., Electronics Park, Syracuse, N. Y.



Whom and What to See at the Radio **Engineering Show**

(Continued from page 168A)

Associated American Winding Machinery, Inc., Booths 4228-4230

750 St. Ann's Ave. New York 56, N.Y.

L. I. Guttman, G. B. Franklin, W. Forster, M. Meier, E. Spiess, G Embree, B. Hammerman meter, E. Spiess, G Embree, B. Hammerman *Fully Automatic Coil Winding Equipment. Miniature, Small, Medium, Large, *Precision Winding With Alternate Spindle. *Perfect Winding, *Dual Spindle Winding, Winding Finest Wires Obtainable For Production or Laboratory. The Most Unique No-Dwell Elec-trical Traverse Ever Developed. No Gears. No Cams

> Associated Testing Labs., Inc. 115 Clinton Road Caldwell, N.J. Booths 3830-3832

Bernard Novack, William Tonkowich, Robert Goldsmith, Frank Keena, Bernard Brodsky, Nelson Burack, Jack Bystrom, Albert F. Erd-man, John E. Stryker, Daniel N. Schochet



Acoustic Noise Test Facility

Manufacturing Division: Manufacturer of envi-ronmental test chambers and equipment for acous-tic noise, saw oth shock', auxiliary vibration (slippery tables)*, temperature-humidity (cascade and liquid CO₂), salt spray (all-Lucite). Testing Divisions: Environmental Testing Divisions, lo-cated at Caldwell N.J. and Orlando, Florida, per-forming environmental and reliability testing.



ACTUAL SIZE

Solid Tantalum, Metallized Mylar, B Mylar, B Dielectric, Electrolytic, Mil-C-25A Paper, Metal-lized Paper, RF Noise Suppression Filters, and Gerunic Capacitors (Skottie Electronics, Inc.).

Atlas E-E Corp., Booth 4222 See: Atlee Corporation

Atlee Corporation, Booth 4222 47 Prospect St. Woburn, Mass.

▲ Allan Q. Mowatt, Dan E. Baker, Frank T. McAvoy, Leverett A. Martel

Trimmer capacitors, heat dissipating tube shield, full contact insert for tube shields, transistor clips and heat sinks, standard and special clips and holders for all types of components.

(Continued on page 174A)

▲ Indicates IRE member. * Indicates new product.

March, 1960

World Radio History

Astron Corp. 255 Grant Ave. East Newark, N.J. Booth 2602

I. I. Ser, ▲ P. M. Maler, R. Black, A. Burton, A. Merola, J. Gordon, R. Mottola, A. Walker, ▲ H. Mutz, J. Barg, L. Busch, A. Gordon, I. Lubin, E. Pataki, R. Heller



Microwave Component News

Lower relay equipment operating costs with new Sylvania Klystrons

Metallurgical and processing improvements mean superior life and performance

Sylvania's research and production capabilities have produced a series of klystrons that promise to surpass earlier types in performance.

Sylvania's klystrons have the following features:

Improved high-temperature glass seal this permits higher bake-out temperatures and gives a lower gas level. The resulting tubes have a life expectancy of 10,000 hours, 2,000 hours longer than competitive types, and better shelf life. This means lower operating costs for relay link equipment.

Purer metals and materials – the premium quality metals used in these tubes, combined with new, exacting processing techniques permit higher bake-out temperatures and result in longer trouble-free operation with low gas levels.

Superior performance – full coverage from 5925 to 8100 mc with 1 watt nominal output power. Most of these tubes have a minimum electronic bandwidth of 28 mc.

Sylvania klystrons will give you added cost savings because of their longer life and fewer early-life failures. Send for the data.



Sylvania Electric Products Inc. Special Tube Operations 500 Evelyn Ave., Mountain View, Calif.



SK-220B, shown approx. 1/3 actual size. Fins facilitate forced-air cooling.



SK-222D, shown approx. 1/2 actual size. Flange connects to heat sink.

Forced air cooled	Conduction cooled	Frequency
SK-220F	SK-222F	5925-6225 mc
SK-220E	SK-222E	6125-6425 mc
SK-220G	SK-222G	6425-6575 mc
SK-220D	SK-222D	6575-6875 mc
SK-220C	SK-222C	6875-7125 mc
SK-220B	SK-222B	7125-7425 mc
SK-220A	SK-222A	7425-7750 mc
SK-220Z	SK-222Z	7750-8100 mc

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Dynacor Square-Loop Tape Cores are manufactured with the high permeability alloys-Grain-Oriented 50-50 Nickel Iron, 4-79 Molybdenum Permalloy, and Grain-Oriented 3% Silicon Iron . . . with fully guaranteed uniformity . . . under rigid standards of control and inspection.

Look to Dynacor for reliable production and swift delivery of your tape core requirements. For your convenience a full line of standard units are stocked for immediate off-the-shelf delivery– Send for bulletins DN 2000, DN 2001, DN 2002.



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Whom and What to See at the Radio Engineering Show

(Continued from page 172A)



Audio Devices, Inc., Booth 2519 444 Madison Ave. New York 22, N.Y. H. Kornbrodt, R. Hickey, J. Puttre, G. Kulper, G. S. Johnson, B. N. Freifeld, R. Haag Type EP AUDIOTAPE—extra precision magnetic recording tape for computers, telemetering, automation, seismography. The complete line of Audiotape for every sound recording need. Audiodiscs, Audiofilm.

Augat Bros., Inc. 31 Perry Ave. Attleboro, Mass. Booth 4504 E. H. Augat, N. F. Damon, C. E. Kiefer, R. C. Hoy, R. S. Laurence, W. H. Sonkin, H. Morrell, D. Sonkin



Component clips, transistor clips, Heat dissipating tube shield (sub-miniature). Crystal socket clip assembly, *sub-miniature relay socket assembly, panel mounting brackets, sub-miniature tube cradle and socket assemblies, potentiometer clamp rings, servo motor clamp rings, tube clamps, custom metal stampings.

Automatic Electric Sales Corp. Northlake, Ill. Booths 1906-1908

R. B. Liepold, T. E. Smith, ▲ V. E. James, J. W. Schaffer, ▲ H. P. Hohberger, ▲ A. T. Brennan, R. O. Cuevas, H. A. Grady, G. W. Downs, J. F. Harm, ▲ L. B. Mitchell, D. C. Leis, J. F. Costello, R. Winthrop



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(Continued on page 176A)





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3

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... cuts Space, Weight, Cost, Power Requirements – Sperry's new STL-222 provides twice the gain of ordinary L-Band tubes—actually takes the place of two tubes in most applications – yet is only 20" long, weighs only 8.5 pounds. This important advantage suits this new CW amplifier and driver perfectly to airborne applications. Its excellent broadband stability recommends it for ground support and airborne radar equipment . . . communications . . . drone applications . . . noise generators . . . switching devices and other L-Band uses.

2

The STL-222 is periodic permanent magnet focused. Its tough metal and ceramic construction provides for high environmental capability, stable operation at high ambient temperatures and under extremes of vibration. This tube also features a high-mu modulating grid and high input-to-output isolation. It is short circuit stable.

The STL-222 is now in production at Sperry, which means lower unit cost and fast delivery schedules. Advanced performance and dependability result from Sperry's long experience in klystron and TWT research, development and production. Write for complete data, outlining the nature of your application.

Specifications

Frequency Range1.0 to 2.	0 kmc ¹
Small-signal gain	db min
Saturated Power Output2	w nom
Beam Voltage	1000 v
Beam Current	35 ma
Grid Bias	35 v
Grid Current	.5 ma
Grid Cut-off Signal	v max
Heater Voltage	.6.3 v
Heater Current	.2 amp
Input-Output Isolation75 a	db min







SPERRY ELECTRONIC TUBE DIVISION, SPERRY RAND CORPORATION. GAINESVILLE. FLORIDA Address inquiries: Gainesville. or Sperry Offices in Brooklyn • Boston • Philadelphia • Chicago • Los Angeles • Montreal • Export Dept.. Great Neck, N.Y.



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Whom and What to See at the Radio Engineering Show

(Continued from page 174A)

Automatic Mfg. Div., Booths 1218-1224 See: General Instrument Corp.

Automatic Metal Products Corp. 315-323 Berry St. Brooklyn 11, N.Y. Booth 1524 M. W. Martin, P. Gilbert, X. B. K. Green, E. Bergenfeld, M. Ross, J. Onore, H. A. Feiner, George Smith



*New Compact Coaxial Switch

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Automatic Seriograph Corp., Booths 1610-1618, 1709-1717 See: Litton Industries, Inc.

Autotronics, Inc. Box 208 Florissant, Mo. Booth 1104 ▲ E. F. White, ▲ F. Haynes, ▲ A. Lee

Sub-Miniature Clutches & Brakes; Low Voltage D.C. Power Supply Lab, Equipment; *Torque Indicators; *New Pancake Clutch/Brake; Speed Changers; Flea Power Clutch.

> Avco Corporation Crosley Division Cincinnati 25, Ohio Booth 3064

R. E. Stockwell

Scale model of FPS-26 Height Finder Radar system developed by and now in production at Avco/Crosley on a prime contract to USAF.

(Continued on page 178A)

▲ Inducates TRE member. * Indicates new product.



WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

SHOW

BOOTH 4240

FEATURES:

- A new low in noise levels down to 0.007 μv providing 0.02 μv full scale sensitivity at minimum bandwidth.
- 5 db steps
- Attenuation range of 85 db
- FOUR regular VSWR scales plus one expanded
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- Continuous gain control over 15 db range
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- Front panel meter monitors bolometer bias current

SPECIFICATIONS:

- Frequency: 1,000 cps; adjustable over a 2% range.
- Sensitivity: $0.02 \ \mu v$ at minimum (4 cps) bandwidth. $0.1 \ \mu v$ at maximum (40 cps) bandwidth.
- Noise Level: 5db below full scale (0.007 μ v at minimum bandwidth).
- Amplifier Q: 250 at 4 cps; 25 at 40 cps.
- Bandwidth: Continuously variable from 4 to 40 cps.
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- Range: 85 db. Input attenuator provides 70 db in 5 db steps. Gain control provides 15 db adjustable. Accuracy ± 0.1 db per 10 db. Maximum cumulative error of ± 0.2 db at 40 cps bandwidth.
- Scale Selector: Expanded, Regular, and Bolometer Current. Meter scale always normalized when switching from scale to scale or from expanded to regular.
- Meter Scales: SWR: 1-4; SWR: 1.8-6; SWR: 3.2-10; SWR: 6-15; Expanded SWR: 1-1.3; db: 0-10; Expanded db: 0-2.3.
- Input Selector: 220,000 ohms; Crystal; Bolometer. Bias provided for high 8.4 ma bolometer or 4.3 ma low current bolometer. Bias adjustable $\pm 15\%$. A bolometer protective circuit permits any switching operation or cable connectdisconnect without damage to bolometer.
- Output: Jack for 1500 ohm recorder, 1 ma full scale deflection.

Input Connector: BNC Jack.

- *Power:* 115/230 v ±10%, 50-60 cps, 40 watts.
- Dimensions: Cabinet: 734" wide, 101/2" high, 11" deep.
- Weight: 14 lbs. net.

See the PRD 277-B Standing Wave Amplifier at the I.R.E. Show— Booths 3602-3604-3606



standing wave amplifier defies comparison



The new portable PRD 277-B Standing Wave Amplifier is designed to meet the present and future needs of microwave test laboratories. Due to its extremely low inherent noise, $0.007 \ \mu v$, weak signals which once were undetectable by conventional instruments can now be measured. Attenuation in 5 db steps combined with 4 VSWR scales and a large meter permit VSWR measurements to be made with maximum resolution and accuracy.

To find out more about the new PRD 277-B Standing Wave Amplifier, contact your local PRD representative or phone, write, or wire:

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Whom and What to See at the Radio Engineering Show

(Continued from page 176.4)

Aveo Corporation Research & Advanced Development Div. 201 Lowell St. Wilmington, Mass. Booth 3065 J. E. Wilder, R. C. Amiot, A.R. Corrado, A.G. Jensen, R. J. Burns, R. P. Conniff, D. G. James, E. D. Kenna, A.B. Slavin, A.R. D. Grange, R. Mack, N. D. Hudgins Live Kerr Cell demonstrates sub-millimicrosecond shuttering capability. Plasma tienerator spray system for deposition of refractories: Rotating Mirror Cametas. Crosley radar shown.

Avion Division, Booth 1113 See: ACF Electronics Division

Avnet Electronics Corp. 70 State St. Westbury, L.L., N.Y. Booth 1103 C. Grey, ▲ R. Hayflick, R. Erhardt, E. Cooney, J. Yonis, J. Nelson, G. Contino, C. Merz, J. Walsh



Automatic Insulation & Circuit Tester

Automatic Insulation and Circuit Tester –Performs automatic production testing of Electrical Components particularly Connectors and Cables for Continuity, Insulation Resistance and High Potential Breakdown– Bendix Connectors, Robertson Splice and Connector Cases, U.S. Semicor Semiconductors Sangamo Capacitors–Vibrex Fasteners

> Axel Brothers, Inc. Electronics Div. 134-20 Jamaica Ave. Jamaica 18, N.Y. Booth 1108

C. Benjamin Axel, R. H. Elkes, M. L. Matnick, C. D. Bitteti, ▲ J. P. O'Donnell, S. Spiegel, S. Zweig, J. Alaimo



Axel Pulse Forming Network

Manufacturers of high-voltage capacitors, low inductance capacitors, pulse forming networks, R, F, suppression filters, special capacitors and related networks.

(Continued on page 182.4)

▲ Indicates IRE member. * Indicates new product.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960




DIRECTIONAL GYRO

This compact new gyro, designed for application in high-performance aircraft and missiles, provides extremely accurate attitude data. Its liquid bubble-type vertical sensing element generates error signals proportional to spin axis displacement from horizontal, while minor wiring modifications permit sensor connection to leveling torquer, completing inner axis leveling loop.

TYPICAL

CHARACTERISTICS # A2215 Environmental Capabilities

Vibration: 5g, 20-1000 cps; 10g, 1000-2000 cps Temperature Range (operative): -54 C to 71 C (non-operative): -65 C to 85 C Altitude: Unlimited Azimuth Pickoff Excitation: 26V, 400 cps, single phase Output (sinusoidal): 11.8V ± 5% max. Error from E.Z.: 10 min max. Motor Excitation: 115V, 400 cps, three phase Speed: 23,500 RPM Power: Starting. 35 watts Running: 7.5 watts Performance Characteristics Drift: 4° hr. max. Leveling Rate: Between 2° and 4° min. Azimuth Torquing Rate: 360° min. (intermittent) 40° min. (continuous)

Write for complete data.

Synchronous Motor





Engineers: Kearfott offers challenging opportunities in advanced component and system development.





VERTICAL GYRO

Kearfott's rugged new vertical gyro, designed for missile application, is a two-degree-of-freedom instrument with 360° of freedom about inner gimbal axis. Self-contained vertical erection system incorporates liquid bubbletype vertical sensing device.

TYPICAL CHARACTERISTICS # B2115 **Environmental Capabilities** Invironmental Capabilities Vibration: 5 g, 20-1000 cps; 10 g, 1000-2000 cps Temperature Range (operative): -54 C to -71 C(non-operative): -65 C to -85 CAltitude: Unlimited Pickoffs Excitation: 26V, 400 cps, single phase Error from E.Z.: 10 min. max. Output Voltage (line to line): 11.8V = 5% max. Motor Excitation: 115V, 400 cps, three phase Power: Starting: 35 watts Running: 7.5 watts Performance Characteristics Repeatability of Established Vertical: To within a cone of half angle equal to 12 minutes of arc Scorsby Drift Rate in 5 Min. Time: Scorsby Drift Rate in 5 Min. Time: 0.3°/min. (average) Erection Rate: Normal: Between 2° and 4° /min. Fast: 80°/min. intermittent, 40°/min. Continuous Physical Features Anisoelastic Drift: 0.08°/min/g² at resonance Weight: 5.5 lbs. (approx.) Mass Unbalance: 0.1°/min 'g Write for complete data.

Rotary Switch

BASIC Building Blocks From Kearfott



FREE GYRO

A highly reliable, twodegree-of-freedom instrument utilizing AC synchro transmitters at each gimbal axis. Designed to operate under the most severe missile conditions, this gyro has AC torquers mounted at each gimbal axis to permit command positioning or slaving of spin axis to desired reference position; each torquer eapable of produeing a precession rate of 360 /minute with 12.5 watts phase power input.

TYPICAL

CHARACTERISTICS = Q2315 Environmental Capabilities Temperature Range: (operative): ---54°C to - 71 C (non-operative): ---65 C to 85 C Altitude: Unlimited Vibration: 10g, 10-2000cps Pickoffs Excitation $\begin{array}{l} 26V, 400 \mbox{ cps, single phase} \\ 0utput (sinusoidal): \\ 11.8V \pm 5\% \mbox{ max.} \\ \mbox{Error from E.Z.: 10 min. max.} \end{array}$ Motor Excitation 115V, 400 cps, three phase Speed: 23,500 RPM Momentum 2.25 x 106 gm cm2/sec Caging and Preset Provision (Electrically energized torquer type) Excitation: 115V max./phase Torquer Constant: 22.8 dyne.cm/Volt2 Performance Characteristics Free Drift: 5°/ minute each axis Runup Time 1 minute max. Torquing Rate: 360 min. (intermittent) 40° min. (continuous)

Write for complete data.



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- 0.01 v/div to 20 v/div.
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Whom and What to See at the Radio **Engineering Show**

(Continued from page 182A)



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(Continued on page 186.4)

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Whom and What to See at the Radio Engineering Show

(Continued from page 184.4)

Barnstead Still & Sterilizer Co. 2 Lanesville Ter. Boston 31. Mass. Booth 4004 A. M. Fulton, D. G. Miller, S. Atkins, V. C. Smith, N. A. Everett, B. M. Greely, E. Morgan



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Basic Products Corp., Booths 2815-2819 See: Sola Electric Co.

Beattie-Coleman, Inc., Booth 3823 1000 N. Olive St. Anaheim, Calif.

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Beckman Instruments, Inc., Helipot Div. & Shockley Transistor Corp., Booths 1201-1205

See: Helipot Div. & Shockley Transistor Corp.

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Inertial guidance and navigation systems and components, all-weather automatic landing system, control and recovery systems for missiles and drones, beacon systems, surveillance and countermeasures systems.

(Continued on page 188A)

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(Continued from page 187.4)

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Plugs, Sockets and Connectors; miniature and subminiature, including coaxial types, Pull range of printed circuit accessories, fuse holders and fuses. Protective devices, noise suppression and filtering equipment, diplexers, triplexers, attenuators, terminal blocks and terminal and nuclear reactor connectors.

Bendix Aviation Corp. Bendix-Pacific Div. 11600 Sherman Way North Hollywood. Calif. Booths 2222-2232, 2329-2331 C. E. Ruckstuhl, E. L. Nolan, D. E. Wassall, John Familetti, R. L. Ramsey, E. W. Copeland, Charles Thomas Development and manufacture of missile system—FM/FM telemetering systems both airborne and ground stations—Sonar and underwater ordnance—digital data transmission, handling and control equipment—airborne radar and radar beacons —sonic and radar altimeters.

Bendix Aviation Corp., Eclipse-Pioneer Division, Booths 2222-2232, 2329-2331 Teterboro, N.J.

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See: M. C. Jones Electronics Co., Inc.

Bendix Aviation Corp., Montrose Division, Booths 2222-2232, 2329-2331 South Montrose, Pa.

M. G. Douglas, & R. F. Henry, S. E. Robinson, D. G. Snyder, R. W. LeGrand, J. H. Speicher, A W V. Fiore, C. C. Honeywell, R. F. Gillen, S. Beers, M. Sanchez, H. Sarrides

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(Continued on page 190A)

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188A

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Sea level (to ground)	1800	3000
Altitude 3.4 in. hg. (adj. terminals)		
50,000 ft	500	800
Altitude 3.4 in. hg. (to ground)	600	900
VOLTAGE RATINGS:		
Seal level (adi, terminals)	550	850
Sea level (to ground)	600	1000
Altitude 3.4 in. hg. (adj. terminals)	2/0	0.00
50,000 ft.	160	250
Altitude 3.4 in. hg. (to ground)	200	300
RECOMMENDED WITHSTANDING VOLTA	GE:	
Seal level (adi. terminals)	. 1200	1500
Sea level (to ground)	. 1300	1600
Altitude 3.4 in. hg. (adj. terminals)		
50,000 ft	. 350	600
Altitude 3.4 in. hg. (to ground)	. 450	700
Current Rating:	1	ampere
Contact Resistance: 0.05	5 ohms N	Naximum
Insulation Resistance: 50,000 M	egohms /	Ninimum
Capacitance: Between one contact and all other conducting parts2	5 mmf N	Naximum
Electrical tests performed in accordance	with EIA	

Standard RS-167.



ACTUAL

SIZE



The socket provides two slots of different widths mating with two

corresponding legs depending from the metal envelope of the tube to index the tube and socket contacts. As a result the tube can be inserted by feel only and it is impossible to insert the tube incorrectly or damage the contacts. The socket saddle provides spring elements that engage with the depending legs of the tube envelope thus grounding the envelope to the panel.

The socket body is of low loss phenolic insulation, Type MFE. The saddle is of cold rolled steel, cadmium plated. The contacts are of copper alloy with cadmium plating.

Although the contact tails are of sub-miniature size, an ample slot is provided for ease of soldering connecting leads.

> The socket fits into a .484 diameter hole with two slots as shown below, and the two legs of the socket that fit into these slots fold over on the under side of the panel, this holds the socket securely in place.

RCA NUVISTOR TUBE

R



Centrally located plants at Chicago, Illinois; Shelbyville, Indiana; City of Industry, California; St. Louis, Missouri

CINCH MANUFACTURING COMPANY 1026 South Homon Ave., Chicogo 24, Illinois Division of United-Carr Fastener Corporation, Boston, Mass.



Whom and What to See at the Radio **Engineering Show**

(Continued from page 188A)

Bendix Aviation Corp. Red Bank Div. Electron Tube Section Eatontown, N.J. Booths 2222-2232, 2329-2331 Booths 2222-2232, 2329-2331 Daniel J. Bell, Donald B. Blanchard, A Joseph F. Bozzelli, James M. Degnan, Rudolph K. Forsman, Thomas B. Mac-Cullough, John H. Moor, R. A. Soerhoff, A Eugene W. Swenarton ELECTRON TUBES: Special Purpose tubes—receiving and transmitting; Gas tubes—toise Sources, spark gans (bath trigger and protective types). Zenon Thy-ratron, Voltage regulators, time totalizers; Microwave tubes—Backward Wave Os-cillators, klystrons, traveling wave tubes, metal ceramic (external anode).

metal ceramic (external anode).

Bendix Aviation Corp. Red Bank Division Semiconductor Products Section 201 Westwood Ave. Long Branch, New Jersey Booths 2222-2232, 2329-2331 ▲ R. R. Mcijer, B. L. Gilluly, R. C. Lancaster, B. D. Gentry, D. I. Snell, ▲ H. Newman, ▲ E. Belmont



Unique DAP (Duffused-Mloy-Power) Transis-tors; High-power Germanium PNP Transistors; Germanium PNP Driver Transistors; Silicon Power Rectifiers; Germanium PNP Power Switching Transistors; Military-type Germanium Power Transistors and Silicon Power Rectifiers.



Benrus Watch Co., Booth 1625 See: Pic Design Corp.

Bergen Laboratories, Inc., Pooth 2840 60 Spruce St.

Paterson 1, N.J.

▲ Max Hoberman, ▲ Tom Bright, ▲ Fred See-kamp, ▲ Walter Katz

Chronistor elapsed time indicator for military and commercial use. G-fuse shock and impact indicator. Driver for silicon controlled reenfiers* for control of large amounts of power from small signals. Electrocap* system of electronic corrosion control.

(Continued on page 195.4)

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World Radio History

March, 1960



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TS 731/URM. Portable, general purpose, field type, rangedized wattmeter used to measure R.F. power out-put of tadlo, radar & natigation equips. A direct reading summation type circuit in which a bolometer acts as one arm of a D.C. Wheatstone bridge, at the same time providing correct resistivity termination for R.F. entering the wattmeter. Power supply: 115/XC, 1 ph, 50-1600 cy, Freq. range: 1000 to 4000 mes with plumbing supplied. VSWR less than 1.3. Gov't cost well over \$1000 in large quantity. Your cost=\$395 ea. Factory new. Well over c. Factory new

ROTARY JOINTS

BAND: Sperry apq36 joint. 1"x1/2", \$15.00 ea. X BAND: Sperry, with $3\frac{1}{2}$ " focal. Length cutler feed dipole $1^{\prime\prime}x\frac{1}{2}$ ", \$15.00 ea.

X BAND: 114"x5%". \$15.00 ea.

X BAND: Azimuth joint 114"x%" APS-15, \$15,00 ea X BAND: Very high speed joint. Dahno Victor, AN/ APS-6, Stainless steel, \$15,00,

S BAND: % coax rotary joint. Gold plated, P/O SCR584 anterna, \$55,00.

L B BAND: 178" coax (no center cond) Western Elec-ic. \$15,00.

TRANSITIONS

RG48 TO % COAX. ADAPTER S Band 1½ x 3" W.G. to RG 44/U coax, \$25,00 ea. RG48 TO TYPE "N" ADAPTER, S band 144" x 3" W.G. \$21,50 ca.

RG51 TO RG52 (1¼" x %" to 1" x ½") smooth taper, \$16 ca.

RG51 TO RG52 FLEX, MAGNETRON eplg. \$15 ea.

CAVITIES

2C40 Lighthouse savity. Tempetature compensated luvar, S band tuneable over range (10cm), Maximum power capability, P/O AN/APW-11A, \$77.50 each

F-28/APN-19 FILTER CAVITY

Jan. spec: Tuncable 2700-2900mc, 1.5db max, loss at etr freq over band, Details: Insertion loss variable. Single tuned filter for freq channelling in radar beacon, invar center tuning conductor ¾ wavelength. New \$37.50 each.

McNALLY Cavity, 2K28 or 707B, Solid coin silver, Tuneable over S band (10cm), \$22,50 each. 2C40 Lighthouse Cavity, Temp. comp. brass, Tuncable over 10cm band, P/O AN/APN-60, \$55,00 each.

1Q22 CAVITY, 9310 mcs. Beacon reference cavity, Mfg. Westinghouse \$17.50 ea.

RADIOSONDE RECORDER: RD-21/AM. Mfg. Times Familie Corp. A complete portable unit with 12" chart roll, \$450. GENERAL RADIO DECADE FREQUENCY CONTROL SYSTEM .1 to 10 nmc. 6 ft rack, 110v 60cy AC input, \$650,

March, 1960

aboratory Standard Potentiometer. A new line of temperatureontrolled 6-dial instruments, reading from 2.101010 volts by vicrovolts, or from .2101010 by .1 microvolts, and calibrated to e useable for linearity to an accuracy of \pm .0002% or better. bsolute guaranteed accuracy of \pm .001% is 10 times better nan any like equipment available commercially today. Excellent

esolution permits use as a aturated standard cell comarator. Available with accesory items such as Four Terninal Type Resistors, Volt atio Boxes, Low Thermal EMF ransfer Switches, Galvanomters, Constant Temperature tandard Cell Enclosures and aturated Standard Cells.



Calibration Console. Accuracy \pm .05% of actual reading DC to 25kc. Ranges .5v. to 1111v. and 1 ma. (2 ma. on AC) to 11.11 amps in decades of .1v. and 1 ma. Resolution \pm .01%. Direct readout in % error as well as in actual values. Console consists of a Hermach type AC-DC thermal transfer standard and a separate DC Calibrator. Complete with \pm .005% stable reference source, high sensitivity galvanometer and self contained controls. Designed for one-man operation! Model LT-PS is a wide range, low distorion (.1%), highly stable AC power supply designed to operate within the full capabilities of the console.



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tadio Frequency 'Self Checking'' tMS Voltmeter

or the certification of /TVM's from DC to 10 Mc. Accuracy \pm .3% of full icale, \pm .2% frequency influence. Ranges .01/.1/1/3 v. Checks its own accuracy igainst a .05% stable inernal standard source. Intrument reading with inanges in frequency is itable and exactly reproducble. Diamond Pivoted of iourse!

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± .01% accurate, ± .005% stable DC reference source! Ranges 1.0000v. and 1.0185v. For use with "null balance" devices such as potentiometers and other infinite impedance comparators. Unaffected by vibration, extremes of temperature, or changes in operating position. Can be short circuited indefinitely without affecting accuracy or life ex.

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formers are wound on high permeability ferrite cup cores and are hermetically sealed in cylindrical brass cases approximately $\frac{1}{2}$ " long with a diameter of 0.4". The units can be obtained with two or three windings and a choice of nine different turns ratios. Connections are provided through pig-tail type leads, $1\frac{7}{8}$ " minimum length.

The transformers are available in three grades designed to meet different environmental specifications: Commercial Grade, operating at temperatures from -25° to $+105^{\circ}$ C; MIL Grade, operating from -55 to $+105^{\circ}$ meeting Grade 5, Class R requirements of MIL-T-27.V and MIL-T-21038; and X-Grade, which meets all MIL Grade requirements and has an increased temperature specification to $+150^{\circ}$ C.

Microwave Attenuators



This new series of low temperature coefficient microwave attenuators is available in 3, 6, 10 and 20 db from **Microwave Control Corp.**, 250 W. 57th St., New York 19, N. Y. The attenuators are available in type N fittings and cover the range of 1 to 10 kmc with low VSWR. The attenuators employ precision vacuum evaporated metal film resistors with a T_c of 280 ppn, resulting in fixed attenuators exceptionally stable from -55° C to $+125^{\circ}$ C.

High Speed Sampling Relay Catalog

A catalog just released by James Electronics Inc., (formerly James Vibrapowr Co.), 4050 North Rockwell St., Chicago 18, Ill., illustrates and gives full technical details of the firm's new line of "Micro-Scan" relays designed for DC, asynchronous and synchronous switching of extremely low microvolt level to moderate level signal circuits such as found in digital, analogue and measurement applications.

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March, 1960

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R SIXTEEN CHANNEL Assemblies with 16 channels. 24 inch paper permits up to 24 channels on one inch centers.

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Write on your company letterhead for Bulletin 891, a 20 page, 2 color brochure giving you complete specifications, application data, etc.

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BMM/B CONSOLE Features horizontal paper travel. Available as Type R or RC.



RC DESK-TYPE CONSOLE Medium gain assembly for computer write-out, telemeter ing, applications with input signal above 10 millivolts

VERSATILITY of assemblies—select the mounting best suited for your use.

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VERSATILITY of writing media—use ink, heat, electric interchangeably on one assembly-select the one most suited for the application.

The Offner Type R Dynograph Assembly is unmatched for sensitivity, accuracy, versatility-we invite you to compare it with any other high speed direct writing oscillograph.



NER ELECTRONICS INC. River Road, Schiller Park, Illinois (Suburb of Chicago) 3912





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(Continued from page 498.4)

Designated series RB-500, the group includes, portable, bench and rack mount models, in general purpose, deviation, sinecosine, binary and automatic stepping types. In all models precise voltage divi-sion is accomplished by means of an adjustable ratio transformer. Unlike resistive dividers, design provides the advantages of high input and low output impedance, preventing loading of the input circuit and minimizing the effect of capacitance between the bridge arms and ground.

All models feature in-line window readout, and provide a range of ratios from +1.111111 to -0.111111. Measurements about zero and unity, consequently, are accomplished without the need for disrupting test set-ups, as required with dividers that do not provide below-zero settings.

Depending upon the particular model, accuracy specifications are 1 ppm to 10 ppm. The unique design features of the instrument provide rated accuracy both above and below zero ratios. Bench and panel mounting types are designed to occupy minumum space and include input and output connections on front and rear, permitting ready integration into consoles, carts and similar larger instrument systems.

IF Preamplifier



A new transistorized IF preamplifier further supplements the standard line of transistorized equipment designed for use in missile, space, and telemetry applications by LEL, Inc., 380 Oak St., Copiague, N.Y. The Model 86 has a bandwidth of 20 megacycles centered at 60 mc and designed to be used with microwave receiver mixers having an IF source impedance of 300 ohms and 18 $\mu\mu$ f. Noise figure is better than 4.25 db. The Model 86 is also available at other center frequencies and for other source impedances.

Insulation Coating

A high temperature, heavy-duty service, insulation and protective coating,



HumiSeal Type 1H34, has been announced by Columbia Technical Corp., 61-02 31st Ave., Woodside 77, N. Y. This coating is practical in many military applications and particularly in those where radioactive environment is involved. A one-component system on silicone resin basis, HumiSeal 1H34 is characterized by 6month long pot life and excellent electrical properties at temperatures above 400° F. HumiSeal 1H34 may be applied by spray, dip or brush. Further data available from the firm.

DC Drive Choppers

Stevens-Arnold, Inc., 7 Elkins St., South Boston 27, Mass., will release at the IRE Show, March 21-24, Booth 2934, a complete line of DC Drive Choppers, both SPDT and DPDT, featuring low noise and a 94-cps chopping rate. Catalog 554 gives all information and prices.



In portable equipment the availability of these choppers makes it possible to build high performance circuits designed around a chopper with a noise level specification of 1 microvolt into 100,000 ohms.

In the non-portable field, the substitution of a dc drive for the conventional ac drive means that the ac drive wiring is removed from the circuitry. The power supplies used with transistorized circuits are well suited to supply the 12 or 24 volts dc required to drive the chopper.

Subminiature Transistor **Pulse Transformers**

The development of the Type BME series of hermetically-sealed subminiature low power pulse transformers for use with transistorized blocking oscillator and interstage coupling circuits, has been announced by Technitrol Engineering Co., 1952 E. Allegheny Ave., Philadelphia 34, Pa.



The Type BME units are available in a range of pulse widths from 0.05 to 5.0 usec at repetition rates up to 10 mc. The trans-

(Continued on page 502A)

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(Continued from page 494A)



tube is frequently run with the printing head at ground potential.

Used with the new Printapix direct writing tube, ordinary paper provides a low cost base material for image rendition. Paper with a glossy surface, commonly used in many printing applications, will provide excellent results. Printing quality can be improved by rendering the opposite side of the paper slightly conductive. Various transparent media, such as glass and thin transparent plastic or commercial sheet polyesters, may be used with Printapix. Dielectric material transport requirements depend on the proposed application.

Image development with the new Litton Printapix direct writing tube is a simple, inexpensive, instantaneous and dry operation. One system employs a developing powder with two components, a toner and a carrier. Agitation of the combination produces a tribo-electric charging. The toner is a finely pigmented plastic material which becomes positively charged, and is thus attracted to the negative charge image on the dielectric material. A typical carrier material is powdered iron.

The developing powder is released as a cloud or fog in close proximity to the charged dielectric surface. Pigmented plastic is attracted to and retained on the charged areas by the coulomb force. The resultant image can subsequently be erased for reuse of the base material and powder, or be permanently fixed by a rapid heat cycle, pressure or other means. Since the resultant image color is determined by the pigment, multicolor reproductions may be obtained by proper development.

AC Ratio Boxes

A new line of precision AC voltage dividers are featured among the instruments and instrument systems demonstrated by **North Atlantic Industries**, Plainview, L.L., N.Y. at the 1960 IRE Show.



(Continued on Juge 500.1)



L'atest in Coil Winding Equipment Company's wide variety of high-speed coil winders for all possible needs is the Model CK, designed to take maximum advantage of the time and money saving principles of automation. The Model CK meets most applications, minimizes maintenance, and cuts down considerably on the need for special-purpose, custom-built equipment.

CK

MODEL

The Model CK features Coil Winding Equipment Company's recently-developed turret transfer. In combination with a suitable winding head, the ingenious turret transfer permits adding — only as needed hopper feed for the coil forms; stripping, cementing, taping and cutting attachments. The Model CK will produce complete bobbin or single-layer coils without operator attention when equipped with a hopper or magazine for the coil forms and appropriate standard attachments, and it will provide stations for finishing as required.

We'll be pleased to send you complete information. Write or phone:

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This is definitely a "Tall Story," the proof of which is at the Martin Company, Denver Division for Missile Testing. Shielding, Inc. was awarded the contract to engineer six gigantic RF enclosures, **100 feet in height**, for the "Titan" missile, because Shielding has come to be recognized as the leader in the shielded enclosure field.

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Shielding stands prepared to solve your RF interference problems, no matter how large or how small these problems may be.

Write for Brochure #851 which outlines Shielding's capabilities.

> See us at Booths 3061 and 3062 at the IRE Show.

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THE INSTITUTE OF RADIO ENGINEERS

l East 79th Street, New York 21, N.Y.



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(Continued from page 492A)

throughout its range, and better than 0.1° at the 90° point. The power oscillator can be used to excite the amplifier or component under test. The standard frequency is 400 cps, however, other frequencies can be supplied.

The phase shifter output may be used for measurement of phase angle, or quadrature voltage components, or may be used as a signal source for the precise excitation of 2-phase devices.

Direct Printing CR Tube

A new Litton cathode ray tube type has been developed for direct electronic printing at high speed on non-sensitized dielectric material, according to an announcement by the display devices department of the **Electron Tube Div.**, Litton Industries, 960 Industrial Rd., San Carlos, Calif. The tube is tradenamed "Printapix".



First showing of two models of the new tube, in a $2\frac{3}{4}''$ printing head width, will be made during the IRE Show. Both models are available for immediate delivery. Tubes with up to 12'' printing head width can be produced to specific order.

This versatile new electronic component is already being incorporated in facsimile, oscillography, address labeling and television type image reproduction equipment. Other applications soon will include high speed computer readout, controlled information storage and erase for military tactical display maps and stock control uses, projection transparency generation, multiple copy reproduction, and simultaneous recording at any number of dispersed stations.

Tubes employing the above techniques, but utilizing much closer spacing of the writing elements, in order to accurately print minute detail, can be furnished for specific application. Element densities up to one million per square inch are feasible.

Operating circuitry and components of the new Litton Printapix tubes are similar to those normally used for display, readout or oscillographic applications. Ordinary television components and techniques are quite satisfactory in many instances. For operating convenience, the

(Continued on page 198.1)

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March, 1960



Above: Nose cone test shape traveling at 5,500 feet per second photographed with Avco Schlieren System at 0.05 μ sec exposure. Inset: Avco Kerr Cell. Permits exposure from 0.005 to 0.1 μ sec; available as an independent module.

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SEE THE NEW HIGH REP-RATE KERR CELL, AVCO BOOTH 3065, IRE CONVENTION, NEW YORK COLISEUM, MARCH 21-24 complete Kerr Cell Shadowgraph and Schlieren photograph systems.

Write for bulletins describing Avco's products and systems in the fields of Hypervelocity Instrumentation, Major Research Facilities and Environmental Test Equipment. Write to: Products and Services Dept., Research and Advanced Development Division, Avco Corporation, 201 Lowell Street, Wilmington, Mass.

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(Continued from page 490A)

switch has been custom designed to be an integral part of the radar packaging. It occupies 43.7 cubic inches, weighs less than 2.5 pounds and consumes 7 watts of power at 100 volts, 400 cps single phase.

Precision Sub-Miniature Fuse

A new development in fuse construction which allows close resistance tolerances, high reliability in fast blowing characteristics, and complete miniaturization for applications where space is at a premium, has been announced by **Littelfuse**, **Inc.**, Des Plaines, Ill. The new subminiature fuse, known as Microfuse, measures



0.205'' diameter $\times 0.070''$ long. The development behind Microfuse is based on the bead type construction pioneered by Littelfuse in their instrument fuses. This construction along with a new type of filament wire permit uniform resistance and blowing characteristics across the range from 1/500 ampere thru 5 amperes at 125 volts. Blowing specifications are: Life-100% of rating; 0-10 seconds-150% of rating. The devices are available either in pigtail variety, which is especially adaptable for soldered connections, or in the plug-in variety, which is designed to plug into a special sub-miniature fuse holder for chassis or printed circuit board mounting.

> Balanced Coaxial Line Duplexer



Bomac Laboratories, Inc., Salem Road, Beverly, Mass., announces a balanced coaxial hybrid and two cavities for "plug-in" cell-type TR tubes. This unit is in $6\frac{1}{6}$ " line and is rated to handle 10 kw of average power. Other units of this type are available in $1\frac{6}{6}$ ", $3\frac{1}{6}$ " and $6\frac{1}{6}$ " coaxial line.

Airborne Mounting Systems

Robinson Technical Products, Inc., designers and manufacturers of all-metal vibration and shock control mounting systems, has developed two types of mountings which successfully isolate airborne communications equipment from high intensity environments.



Suitable for either military or commercial aircraft, these two designs together compose 55 variations, and are available in many ATR sizes, a wide variety of load ranges, and numerous DPA and DPD connector arrangements.

Model 2310 is a center-of-gravity system designed to accommodate the smaller ATR configurations. The single stage mounting base requires only four prespaced mounting holes, making misalignment impossible. Natural frequency is in the 6–10 cps range with a transmissibility at resonance of less than 5.

Model 2311 is a base type system for larger ATR equipments. Possibility for misalignment during installation is likewise eliminated through the use of four pre-spaced mounting holes.

Each of these mounting systems incorporates MET-L-FLEX stainless steel resilient elements, and each has been designed to meet specifications ARINC 404 and MIL-C-172B.

For further information, write to the firm.

Oscillator/Phase Shifter

The Industrial Test Equipment Co., 55 E. 11th St., New York, N. Y., has developed a new instrument for the precision measurement of phase angles in the vicinity of 90°. The instrument consists of an 8 watt power oscillator and a precision variable phase shifter having a range of $90^{\circ} \pm 10^{\circ}$. The resolution and incremental



accuracy of the phase shifter is 0.05°. The absolute accuracy is better than 0.25°

(Continued on page 494A)

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- INDUCTANCE ADJUSTABLE: ±20%
- ENVIRONMENTAL: Encapsulated in epoxy resin for protection against climatic and mechanical conditions.







The T44VIC is one of several models now available. This noise source is housed in a rigid coaxial line and provides an excess noise ratio of 18.5 ± 0.3 db. The T44VIC is suitable for CW or pulse operation under typical adverse military environments, and will meet all applicable military specifications for shock, vibration, temperature and humidity. The T44VIC will fire and operate at conventional power supply voltages and will operate reliably for a lifetime in excess of 2000 hours.

Two additional models, the T44V2C and the T44V3C, are also available. The T44V2C features a replaceable gas tube element as well as a higher noise output of 21.0 db. The T44V3C also features the replaceable gas tube element, at a noise output of 18.5 db.

For special applications as to frequency or noise output, Tucor can provide a noise source tailored to your requirements. Coaxial versions of the T44V series are available at frequencies up to 1000 mc.

Sampling Switch

Instrument Development Laboratories,

Inc., a subsidiary of Royal McBee Corp., 67 Mechanic St., Attleboro, Mass., announces it has, within 4 months, designed and produced a new 2-pole, 60 position, motor-driven, low-level sampling switch which has run continuously at 60 rps for



more than 200 hours without contact bounce or signal contamination. This switch has performed satisfactorily while undergoing missile vibration testing of 20 to 3000 cps at 35 "G's" for 35 minutes per asis. Designed for application to an Area Defense Missile Guidance Radar System, this switch samples Doppler velocity data for range-rating purposes. With the cooperation of the systems contractor, this

(Continued on page 492.A)





These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

(Continued from page 486.4)

of **Polytechnic Research & Development Co., Inc.,** 202 Tillary St., Brooklyn 1, N. Y., the report describes the theory and operation of both M and O type tubes. The relationship between power output, frequency shift, and delay line current stability is graphically presented as a function of various electrode voltages.

The PRD 813 Universal BWO/TWT Power Supply is discussed in detail and a partial listing of the Backward Wave Oscillator tubes that the supply can operate is shown.

Copies of the Report are offered by the firm on request.

VHF Admittance Bridge

A miniature thermistor element working in a servo feedback system is used as a conductance standard in a VHF Bridge Model 978 manufactured by **Marconi Instruments**, 111 Cedar Lane, Englewood, N. J. This approach is said to guarantee measurement accuracy of 2% to 300 mc. Capacitance and conductance range is $\pm 40 \ \mu\mu$ f and 0–50 millimhos. (Hlustrated are: Admittance Bridge Model 978 and Bridge Source and Detector.)



Two terminal measurements can be made on RF components, semi-conductors, transmission lines, and so forth. The voltage applied to the component under test is small, seldom more than 50 my.

Separate bridge sources and detectors are available for operating the bridge, although any signal generator and receiver covering the frequency range can be used.

For detailed information write to the firm.

Gas Tube Noise Sources

Supplementing the recently announced development and production of compact gas tube noise sources for ground and airborne microwave system applications, **Tucor, Inc., 18** Marshall St., South Norfolk, Com., introduced additional co-axial models for the 200–250 mc frequency range. Available at the T44V series, these new noise sources feature high excess noise outputs. They are suitable for test and calibration of all types of microwave and communications components and systems.

(Continued on page 490.1)



Model 720 events-per-unit-time counter is specially designed for industrial and laboratory uses where unusual flexibility is required. The unique modular construction permits the measurement of virtually any physical variable, such as rate of flow, RPM, frequency or time interval. NIXIE in-line readout is easy to read and reduces operator fatigue and the chance for error. Simplified construction permits operation by relatively unskilled personnel. Low Cost starts at \$475 for the 3-decade model. Send for complete technical literature.

DECADES: FREQUENCY RANGE: SENSITIVITY: ACCURACY: DISPLAY TIME: SIZE: WEIGHT:

3, 4, 5 or 6 0 to 120,000 pulses/sec. 50 Millivolts RMS \pm one count time base error 0.2 to 6 sec. or infinite $19'' \times 51/4'' \times 12''$ 25 lbs.

ERIE-PACIFIC manufactures a complete line of digital counting and timing instruments and systems for military or commercial use.



ERIE PACIFIC DIVISION ERIE RESISTOR CORPORATION 12932 S. Weber Way, Hawthorne, California Phone: ORegon 8-5418

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MARCH 21, 22, 23, 24

The Institute of Radio Engineers 1 East 79th St., New York 21, N. Y.

IRE CONVENTION

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March, 1960

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STEMCO THERMOSTATS

RANK FIRST IN PRECISION TEMPERATURE CONTROL

In today's military and commercial projects, you can't afford to overlook any one of these important areas: Reliability, Size, Availability, Economy.

And because Stevens is in production now on the largest number of different types and styles of bimetal thermostats, all these advantages are yours automatically when you specify Stemco thermostats.

1st in Reliability. Proven designs, latest production techniques, most stringent inspection procedures.

1st in Size. Stemco thermostats score in compactness and lightness without sacrificing performance.

1st in Availability. Tooling for most types is in existence. Flexibility of design cuts lead time on other types.

1st in Economy. Mass production of many standard Stemco types with hundreds of terminal arrangements and mounting brackets cuts your costs.

*Refer to Guide 400EO for U.L. and C.S.A. approved ratings.





TYPE A* semi-enclosed. Bimetal disc type snap action thermostats; give fast response to temperature changes. Can be made to open on rise or close on rise. Single-throw with double make and break contacts. Operation from -20 to 300°F. Lower or higher temperatures on special order. Average non-inductive rating 13.3 amps, 120 VAC; 4 amps, 230 VAC and 28 VDC. Various mountings and terminals available. Bulletin 3000.

TYPE A hermetically sealed. Electrically similar to semi-enclosed Type A. Various mountings, including brackets, available. Bulletin 3000.

TYPE MX hermetically sealed. Snap acting bimetal disc type units to open on temperature rise. 2 to 6°F differentials as standard. 1 to 4°F differentials available on special order. Depending on duty cycle, normal rating 3 amps, 115 VAC and 28 VDC for 250,000 cycles. Various terminals, mountings and brackets available. Bulletin 6100.

TYPE MX semi-enclosed. Construction and rating similar to MX hermetically sealed type. Bulletin 6100.

TYPE M hermetically sealed. Bimetal disc type, snap acting thermostats. Also available in semi-enclosed. Operation from -20 to 300°F. Lower and higher temperatures available on special order. Depending on application, rated non-inductive 10 amps, 120 VAC; 3 amps, 28 VDC. Various terminals, wire leads and brackets available. Bulletin 6000.

TYPE C hermetically sealed. Also semi-enclosed styles. Small, positive acting with electrically independent bimetal strip for operation from -10 to 300°F. Rated at approximately 3 amps, depending on application. Hermetically sealed type can be furnished as double thermostat "alarm" type. Various terminals and mount-ings. Bulletin 5000.





STEVENS manufacturing company, inc. P.O. Box 1007, Mansfield, Ohio



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These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

(Continued from page 484A)

Bandpass Amplifier-Multicoupler

A new distributed bandpass amplifier which provides optimum low noise capabilities and a flat frequency response of from 30 to 300 mc was introduced by **HRB-Singer, Inc.,** State College, Pa., a unit of The Singer Military Products Division.



The 30-300 mc range is obtained from four separate outputs with a minimum input-to-output isolation of 60 db and an average of 30 db isolation between any output.

The HRB-Singer Amplifier-Multicoupler (Model 330-M4) is basically a distributed bandpass amplifier coupled to parallel grounded grid amplifier stages having individual outputs. The amplifier includes an input network which makes possible the optimum low noise capabilities and flat response.

A feature is the parallel connection of the amplifier's six line tubes. With this innovation, failure of several tubes will not seriously lessen the instrument's performance.

The amplifier, constructed with subminiature components and printed circuit techniques, presents a lightweight, rugged package.

The multicoupler case is designed for use in a rack panel installation or for bench experimental testing purposes. The units are shipped ready for use in the rack panel installation and can be removed for bench use by turning the fasteners on the front panels. Identical panel markings are on the sub-panel.

Other specifications of Model 330-M4 are: RF gain 6 db; noise figure 8 db average; input impedance 50 ohms nominal; output impedance 50 ohms nominal; power requirement 117 volts at 1.5 amperes; size $5\frac{1}{4} \times 5\frac{1}{4}$ inches with standard 19inches relay rack panel; size of power supply $8\frac{3}{4} \times 9\frac{1}{4}$ inches with standard 10inches relay rack panel.

BWO Power Requirements

The latest issue of "PRD Reports," Vol. 6, No. 4, entitled "Power Supply Requirements of BWO Tubes" discusses the voltage and modulation requirements needed to power backward wave oscillators.

Authored jointly by Stanley J. Blanchard and Martin J. Blickstein, of the staff

(Continued on page 489A)



DAPON[®] resin (diallyl phthalate prepolymer), a preferred and frequently the only acceptable material in military and space technology applications, is equally well suited to a wide range of industrial uses. Molded DAPON resin parts are widely used today by the leading companies in the electronics industry.

Electrical parts molded with DAPON resin are unexcelled for retention of dielectric strength, insulation resistance and arc resistance even under conditions of high humidity and elevated temperatures. DAPON resin's unusual stability eliminates breakdown and attendant corrosion of metallic electronic components. And the exceptional dimensional stability of DAPON resin permits molding of complex parts with multiple inserts as well as close tolerances without machining. For use at higher temperatures another form of DAPON, named DAPON M resin (diallyl isophthalate prepolymer), is also available.

FMC manufactures the basic DAPON resin only and does not supply finished molding compounds. DAPON resin molding compounds are available from :

ACME RESIN CORPORATION 1401 Circle Avenue Forest Park, Illinois	as	Acme Diallyl Phthalate Molding Compounds
DUREZ PLASTICS DIVISION Hooker Chemical Corp. North Tonawanda, N. Y.	as	Durez Diałłył Phthalate Molding Compounds
MESA PLASTICS COMPANY 11751 Mississippi Avenue Los Angeles, California	as and	Diall® (Diallyl Phthalate) Molding Compounds Diall® (Diallyl Isophthalate) Molding Compounds
ROGERS CORPORATION Rogers, Connecticut	as	Rogers Diallyl Phthalate Molding Compounds

For complete information about DAPON resin's outstanding electrical properties and how it can help solve your problems, write to FMC Chemicals and Plastics Division.

Putting Ideas to Work FOOD MACHINERY AND CHEMICAL CORPORATION Chemicals and Plastics Division 161 EAST 42ND STREET, NEW YORK 17, N.Y.

485A




ELECTRONIC

ADJUSTABLE-SPEED DRIVES

- SPEED RANGE Infinitely adjustable from less than 36 rpm to more than 3600 rpm while delivering full roted torque. Continuous duty rating at all speeds.
- REGULATION Both line and load regulation is better than 1/2 of 1% of roted speed.
- HORSEPOWER Various models from ¾ hp down to 1/200 hp. Motors of ¼ hp and larger are totally enclosed.
- REMOTE CONTROL A 10-turn potentiometer provides precise adjustment ot ony convenient location.
- GEARED MOTORS Motors are available with integrol gear reducers.
- O BRAKING-REVERSING Relay-controlled braking and reversing models available.
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- O OTHER MODELS Servo-Tek manufactures drives with silicon rectifiers and adjustable autotransformers, as well as other thyratron drives with less exacting specifications. Write for information including details of your propased use.

IMMEDIATE DELIVERY





These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

(C ntinued is in page 482.4)

Decimal Electronic Switch

The Beam-X, a new decimal electronic switch, is expected to effect a major change in basic electronic design logic from binary to decimal systems according to **Burroughs Corp., Electronic Tube Div.,** Plainfield,



N. J. The Beam-X switch uses small rod magnets within a vacuum to control the position of an electron beam to any one of ten output positions. The result is a decimal switch so reduced in size, weight, cost and power as to outperform all existing vacuum, magnetic, and solid state devices in multiposition switching, counting, distributing, multiplexing, and allied operations. In a typical ten-position switching application, the decimal switch eliminates the 90 transistors, diodes, and resistors which must be used with binary logic to achieve the same results.

The Beam-X switch type BX-1000 is the first in a new series. Though functionally similar to its predecessor, the Beam Switching Tube, its radical design makes it a completely new device. The BX-1000 is 10 times lighter ($1\frac{1}{2}$ ounces), 5 times smaller (3 cubic inches) and $\frac{1}{2}$ the price (less than \$25.00 in small quantities).

The BX-1000 has useful constant outputs, positive switching elements, and memory in each of its ten positions, may remain stationary indefinitely or switch at speeds exceeding 10 megacycles either sequentially or at random, be interconnected as a distributor of any number of positions less or greater than 10, and be preset to any position and reset in less than a microsecond. Operating flexibly and efficiently with respect to B voltages, it can be utilized equally well in high or low voltage systems. In vacuum tube circuits, outputs as high as 200 volts can be obtained while in transistorized systems it can be operated by 12 volt signals directly from the solid state circuitry. Ruggedly constructed to withstand shock and vibration, and insensitive to temperature extremes, the new Beam-X switch is suited for applications in ground support equipment, missiles, aircraft and space technology, and in commercial and industrial products.

Deliveries will start in February with production quantities being available in March.

WHEN WRITING TO ADVERTISERS PLEASE MENTION --- PROCEEDINGS OF THE IRE

Silicon Rectifiers

Silicon Rectifiers with ratings of 20–35 amperes, 60–600 P I V which exhibit more stable characteristics at high temperature than before are now available from **Dallons Semiconductors, A Division of Dallons Laboratories, Inc.,** 5066 Santa Monica Blvd., Los Angeles, Calif.



The units contain solders within their construction which have a melting point in excess of 600°C. The 14″ stud construction houses a pure silver, heavy spring lead anode assuring ruggedness and high resistance to shock and vibration. Electrical specifications show that these units have less than 5 ma reverse current and the maximum forward drop voltage at a test temperature of 25°C at 20 amperes dc is 0.65 volt.

These units were designed for power supply and magnetic amplifier applications demanding high reliability.

(Centinned on page 486.1)



duPont's trade name for Polytetrafluoreothylene

March, 1960



Horizons Unlimited

Development of a high-performance inertial guidance system of unprecedented performance for long-range guided missiles, satellites and space vehicles has been announced by Bell Aircraft Corporation's Avionics Division.

Bell Avionic's engineers describe the highly-classified system as "the most successful and reliable of any new inertial instrumentation concepts so far tested."

The system was developed under the direction of Dr. Helmut W. Schlitt, recognized within the industry as an outstanding authority in the field of inertial guidance.

The new system has undergone extensive flight tests at the Niagara Falls, N. Y., Municipal Airport. Some of its components already are being used in guided missiles.

For more information about Bell Avionics INERTIAL GUIDANCE SYSTEMS AND COMPONENTS, as well as Battlefield Monitoring Systems, All-Weather Automatic Aircraft Landing Systems, Secure Data Link Systems and many others, you are urged to talk to Avionics Division engineers in Booths 3822 and 3824 during the I. R. E. Show.

ELECTRONICS ENGINEERS

Bell Aircraft's Avionics Division will conduct interviews in New York City during the I.R.E. Radio Engineering Show March 21 through March 24 for urgently needed competent, qualified engineers in the following categories:

Dr. Helmut W. Schlitt,

manager of the Avionics Division's Inertial Development Laboratories, who has directed the development of the high-performance inertial guidance system.

Dr. Schlitt earned his doctor's degree at the Technical University in Darmstadt, Germany, and after coming to the United States he was employed by the U.S. Army at White Sands and Huntsville before joining Bell Aircraft in 1952.

SEE YOU AT THE

IRE SHOW

3824



Electronics Engineers to design and develop transistor circuits for digital systems.

Electronics Engineers to design digital and data handling systems in connection with inertial navigation equipment.

Electronics Engineers to analyze digital computers and systems.

Electronics Engineers to design complex transistor circuits operaring over a large temperature range in inertial guidance systems.

Electronics Engineers to design very high frequency receivers and multi-stage transmitters.

Dynamics Engineers to conduct simulation studies of compatability of aircraft with automatic all-weather landing systems.

Marketing Engineers for complex electronics systems involving radar, airborne communications and flight instrumentation.

For a personal, confidential interview at the Savoy Hilton Hotel, Fifth Avenue at 58th Street, while you are attending the show, telephone CI 7-2805 and ask for Mr. George Klock, director of engineering employment.



World Radio History



AN IMPORTANT ANNOUNCEMENT TO ALL IRE MEMBERS AND SUBSCRIBERS

The IRE Professional Group on Antennas and Propagation has just published the "Proceedings of the URSI International Symposium on Electromagnetic Theory," held at the University of Toronto, Canada, on June 15-20, 1959, as a special supplement to Volume AP-7 (1959) of the IRE TRANSACTIONS on Antennas and Propagation.

Those who registered at the Toronto Symposium will automatically receive one copy as a part of their symposium registration fee. PGAP members and others may obtain a copy by ordering at the rates indicated below. There will be no free distribution because of the special nature and large size of the supplement.

This imposing 400-page volume, comprising invited papers by 54 of the world's leading authorities, promises to be one of the outstanding reference works in its field. The subjects covered include Diffraction and Scattering Theory, Radio Telescopes, Surface Waves, Boundary Value Problems, Propagation of Waves, and Antennas. The complete program may be found on page 18A of the June, 1959 issue of the PROCEED-INGS OF THE IRE.

IRE members and subscribers are urged to order their copies now by returning the form below to the IRE, accompanied by remittance made payable to The Institute of Radio Engineers.

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These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

(Continued from page 480A)

Coaxial Attenuator

Microlab, 570 W. Mount Pleasant Ave., Livingston, N. J., has just introduced a new line of coaxial attenuators which are shorter and lighter than those previously available. This new group of attenuators, known as the AC series, operates from 4000 to 12,000 mc and are as short as 2.5 inches.



The AC series of attenuators are constructed as coaxial lines with lossy center conductors. This method is said to result in a unit with VSWR and attenuation characteristics superior at microwave frequencies to those which can be obtained with lumped constant techniques. The variation of attenuation versus frequency is negligible across the entire band.

These attenuators are supplied from stock with N, BNC, TNC, C, or HN connectors at values of 3, 6, 10 and 20 db. Other values can be supplied on request with delivery generally within two weeks. The maximum input power is 2 watts average, 3 kilowatts peak. They are designed to meet the environmental conditions of MIL-E-5272B and other similar specifications. Their over-all length is approximately 2.6 inches for the lower values, increasing to 3.5 inches for the higher values.

The AC series attenuators are available for immediate delivery from stock and are priced at \$30 to \$45 depending upon the attenuation value and connector type. Further information can be obtained from the firm.

Parts Catalog

Sterling Precision Corp., 17 Matinecock Ave., Port Washington, L.I., N. Y., announces publication of its new catalog #61 consisting of 512 pages with 20,000 items listed from stock. Prices are included.

In its pages are a wide range of gear heads, speed reducers, differentials, precision gears, couplings, shafting, electronic hardware, and so forth

(Continued on page 484A)

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

Skydyne Inc

TRANSIT CASES MINIMUM WEIGHT MAXIMUM STRENGTH



"SET" FOR SERVICE

ON LAND — SEA — AIR

Electronic equipment for ground support of missile or conventional artillery . . field radar . . . tracking systems . . . test systems . . . reconnaissance . . . communications **must**. be in instant operating order . . . whenever and wherever required.

These units . . in a combined operational and transportation enclosure by Skydyne . . withstand the hazards of transportation shock, vibration, weather conditions, dust, sand . . . because the "package" is scientifically engineered for both equipment and service . . designed and built by Skydyne, and backed by over twenty years experience in transit packaging of instruments and equipment for industry and the government.

Skydyne Transit Cases feature

A completely new basic concept of construction for strength, durability and performance, engineered design for maximum protection to meet your system requirements.

When pounds and protection count .. count on Skydyne for transit cases. Write for complete details





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Glass-Tite's story is simple and successful. They make a custom seal to your most minute specifications or they supply standard glass-to-metal seals — either way they make only one quality . . . THE BEST. No design requirement is too tough, no reasonable delivery date is impossible.

Specify Glass-Tite in Kovar and compression seals, diodes and rectifier enclosures for the semi- conductor industry and all electronic applications.

Write Dept. 729 for Interature and send details of your design requirements.

I.R.E. BOOTH 1109

563-9





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(Continuea from page 478A)

Silicon Mesa Transistors

Hoffman Electronics Corp., 3761 S. Hill St., Los Angeles 7, Calif., announced it is now in production on two newlydeveloped silicon mesa transistors.

Because of design features, resulting in an unusually high small signal current gain, either of the devices will replace up to three transistors of the same classification in many circuit applications, the firm said.



The new diffused junction, drift field mesa transistors (JEDEC No.'s 2N696 and 2N697), are designed for use as high speed switching units operating at medium power levels and as very high frequency amplifiers.

Hoffman's U-shaped base-emitter configuration is said to allow utilization of virtually all the transistor's emitter area. The minimum high frequency gain at high currents is 6 or more at 20 mc, nearly three times the 2.5 gain of comparable units. This higher gain is due to tighter control of the base width in fabrication.

The new transistors are capable of useful current gains at 40 mc, indicating efficient operation in the ultra high frequency band when operated in a grounded base configuration.

Both transistors are basically control devices for small to large signal switching amplification. The only difference between the two is a higher dc pulse current gain in the 2N697. This measures a minimum of 40 and a maximum of 120 compared to a minimum of 20 and a maximum of 60 in the 2N696.

The company's standards require preaging the transistors at 300°C before hermetically sealing them in a controlled inert gas atmosphere to stabilize the electrical parameters.

Total power dissipation of the two transistors is two watts at 25°C case temperature.

The transistors have a maximum collector-base voltage of 60 volts, collectoremitter voltage of 40 volts and an emitterbase voltage of 5 volts.

In saturation, with a base current of 15 unilliam peres and a collector current of 150 milliamperes, the emitter-base voltage is less than 1.3 volts and the collector-emitter voltage is less than 1.5 volts.

Price of the 2N696 and the 2N697 is \$28.50 each in quantities of from one to 99 units, and \$19 each in quantities of 100–999 units.

24 Hour Clocks

In addition to the direct reading digital PlanetGear clocks, **Haydon Instrument Co.**, 165 W. Liberty St., Waterbury 20, Conn., now introduces Model 4003, the first in a series of conventional dial type 24 hour clocks. This is an easy to read, precision built, heavy duty clock made especially for flush panel mounting in an instrument board.



The case and mounting flange are black anodized aluminum. Model 4003 may also be furnished in gray or other colors to match existing equipment. The main figures are $\frac{1}{4}$ " high conforming to MHL spec 33558 ASG, and are painted white against a black background. The hour, minute and sweep second hand are also white. The outside diameter of the flange is $6\frac{7}{16}$ "; bolt circle is 6" and dial face 5" in diameter. The over-all depth of case is $3\frac{3}{4}$ ".

Synchronous timing motors available are 50, 60, 400 and 500 cps 110 volts.

The clock has a two year guarantee against defects in material and workmanship.

List price for 50 or 60 cps is \$88.00 with discounts for quantity. Delivery about two weeks.

List price for 400 or 500 cps is \$120.00 with discounts for quantity. Delivery about four weeks.

SWIRECO

The 12th Annual SWIRECO will feature a National Medical Electronics Conference under the auspices of the national executive committee of the IRE Professional Group on Medical Electronics. Four simultaneous sessions will be held at the Shanrock-Hilton Hotel in Houston on April 20 to 22, 1960.

There will be two general sessions, one medical session, and one student session. Ninety per cent of the available exhibit space has been sold. There will be a full program of activities for the ladies. An attendance of 3,000 is typical for SWIRECO conventions.

(Continued on page 482A)

"^s world of silicon selenium

4

Milliwatts to Kilowatts of DC power



LOW POWER

ITT LOW POWER DIFFUSED JUNCTION SILICON RECTI-FIERS. DERBY AND AXIAL LEAD TOP.HAT TYPES; SERIES C AND SERIES R - UP TO 15 AM. PERES. 50 TO 800 VOLTS. PIV.



SELENIUM MINIATURE RECTIFIERS

LOW CAPACITY, VERY HIGH VOLTAGE, VERY LOW CURRENT FOR APPLICATIONS IN DUST PRECIPITATION, ZERO GRAPHIC EQUIPMENT, AND HIGH VOLT-AGE POWER SUPPLY FOR CATHODE RAY TUBES SMALL - RELIABLE - REPLACE HIGH VOLTAGE LOW CURRENT TUBES 1 8 CELL, WITH OR WITHOUT PIGTAILS



MEDIUM POWER

ITT MEDIUM POWER DIFFUSED JUNCTION STUD TYPE RECTI-FIERS STANDARD JEDEC PACKAGES SERIES K, U. AND D 6 AMPERES TO 30 AMPERES, 50 TO 800 VOLTS PIV.

(Outboard Motors)

ENGINE GENERATOR RECTIFIERS

SMALL, RUGGED, LIGHTWEIGHT RECTIFIERS TO SUPPLY DC POWER FROM GASOLINE OR DIESEL ENGINE GENERATORS DRIVEN BY A SIMPLE FLY

WHEEL, DESIGNED FOR OUT DOOR USE, UNAFFECTED BY MOISTURE OR SALT SPRAY

CONDITIONS.



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ITT DIFFUSED JUNCTION ZENER VOLTAGE REGULATOR DIODES SERIES Z. 750 MILLIWATTS TO 10 WATTS - OVER 100 VOLT-AGE RATINGS, AXIAL LEAD AND STUD MOUNTING TYPES.



AUTOMOTIVE RECTIFIERS

HIGH CURRENT RECTIFIERS TO SUPPLY THE HEAVY ELECTRI-CAL DEMANDS OF 2-WAY RA-DIO COMMUNICATIONS IN AU-TOMOBILES, BUSSES, TRUCKS AND OTHER VEHICLES.



PRINTED CIRCUIT

ITT PRINTED CIRCUIT DIF. FUSED JUNCTION SILICON RECTIFIER ASSEMBLIES, SIN-GLE PHASE AND THREE PHASE UP TO 4 5 AMPERES, 50 TO 800 VOLTS PIV.



INDUSTRIAL RECTIFIERS

PROVIDES THE LARGEST ITT. LINE OF CUSTOM-BUILT SE-LENIUM RECTIFIERS IN THE WORLD TO SUPPLY LARGE AND SMALL POWER REQUIREMENTS FOR EVERY INDUSTRIAL APPLI-CATION INVOLVING AC-DC CON-VERSION AT TEMPERATURES UP TO 125°C AMBIENT.

See us at IRE Booth #2510-2520-2615-2625

Write today for startling information on our newly developed, compact, lightweight, high voltage selenium rectifiers which supply a few milliamperes at thousands of volts.

Components Division

INTERNATIONAL TELEPHONE AND TELEGRAPH CORPORATION P. O. 80X 412 CLIFTON. N. J., U.S.A.

World Radio History





Because of the revolutionary bifilar frictionless (no pivot) movement, plus weightless light-beam pointer, GREIBACH PRECISION METERS withstand 100,000% overload surges. Then, for extraordinary overload risk applications, a special built-in Protective Circuit takes up to 125,000,000% overload surges without impairment.



Only GREIBACH offers such overload immunity, along with: Sensitivity down to 0.2 microampere full scale. Accuracy better than 1/4 of 1%. Energy dissipation as low as 4 x 10-1" watt. Permanent reliability. Mechanical ruggedness withstanding up to 500 G's shock.

GREIBACH PRECISION METERS are available in portable, bench and panel models with wide selection of ranges even up to 23 ranges in one meter: e.g., .2 / .5 / 1 / 2 / 5 / 10 / 20 / 50 / 100 / 200 / 500µA / 1 / 2 / 5 / 10 / 20 / 50 / 100 / 200 / 500MA / 1 / 2 / 5-AMP.

Verify these extraordinary advantages by seeing actual demonstrations arranged upon request.



FRICTIONLESS BIFILAR SUSPENSION MOVEMENT

See us at the IRE Booth #3924





These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your IRE affiliation.

(Continued from page 60A)

Differential DC Amplifier

A new transistorized differential dc amplifier, featuring a true floating input and stability of 0.05%, is offered by Neff Instrument Corp., 2211 E. Foothill Blvd., Pasadena, Calif.



In the new Type 1-102, ground loop or circulating current problems in data processing systems have been eliminated through transformer isolation design. The transistorized amplifier operates with both input and output isolated from each other and from ground. Thus, strain gage and thermocouple data, or data travelling over long signal lines, cannot be destroyed through accidental voltage drops.

The heart of the amplifier is a maintenance-free shielded input module which guarantees high common-mode rejection (100,000,000 to 1). Gain is variable in steps of 10, 20, 30, 50, 100, 200, 300 and 500.

The 1-102 may be obtained in either single unit cabinets or with six amplifiers in a 19" rack module. The rack module features a self-contained isolated power supply operating the six amplifiers and an integral blower unit.

For complete information write to the firm.

Single Phase Inverter



Temco Electronics, P.O. Box 6191, Dallas 22, Texas, announced a new single phase inverter designed specifically to supply 400 cps power to rate gyro packages. The inverter is suitable for any application requiring small quantities of 26±1.0% volt ac power. Occupying only 20 cubic inches, this unit is capable of

delivering 20 watts at 400 cps $\pm 0.1\%$ with an input voltage of 28 ± 4 volts. Distortion is less than 4%. Efficiency exceeds 60% at full load. The rate gyro single phase inverter meets or exceeds all applicable portions of MHL-E-5272.

VHF Preamplifier

Community Engineering Corp., P.O. Box 824, State College, Pa., has designed for use in the 50 to 200 mc range, the Model 1001 VHF Preamplifier with a noise figure of better than 3db at 85 mc and 4.5 db at 200 mc with a nominal gain of 30 db. Unit is fixed tuned to required frequency, band width is 10 megacycles.



The circuit employs a GE subminiature ceramic planar triode, 7077, feeding into a 6AM4, followed by two stages of stagger tuned 5654's.

Amplifier sub-assembly and integral power supply are mounted in light weight, solidly constructed chassis, on a standard size rack panel protected by an easily removable dust cover. Input and output impedance is 50 ohms. Standard type N connectors are used.

Other preamplifiers are available covering frequencies up to 900 mc, also, wide band distributed amplifiers. CECO amplifiers can be modified for special application. Weatherproof enclosure available for installations requiring such protection. For further information, please contact the firm.

Coaxial Attenuator

This variable coaxial attenuator produced by Merrimac Research and Development, Inc., 517 Lyons Ave., Irvington 11, N. J., is the first of a new series of wideband coaxial variable attenuators which have flat attenuation vs. frequency characteristics and zero insertion loss.



The unit shown in the attached photograph of model AE-6 has the following characteristics: Frequency range, 4-7 kmc; Insertion loss, less than 0.5 db; Attenuation variation vs frequency, less than ±5% in db; Power handling, 4 watts average; VSWR, 1.5 maximum.

These new attenuators provide up to 40 db of attenuation over the above frequency range and special variations with up to 100 db of attenuation. Other types of coaxial attenuators can be provided.

(Continued on base 480.4)



DIFFUSED SILICON PNPN CONTROLLED RECTIFIER

A three-junction, three-terminal device for use in power control and in switching applications requiring up to 16 amps., D.C. In the reverse direction (anode negative) it will block current up to its rated PIV, while in the forward direction (anode positive) it will block up to its minimum breakover voltage, at which point it will quickly switch to the high conduction state. It may also be turned on when an appropriate voltage is impressed between gate and cathode. In this latter respect it is analogous to a thyratron. In the "on" conduction state, the forward voltage drop is essentially that of a standard silicon diode. Tentative specifications are as follows:

MAXIMUM RATINGS

Peak inverse voltage (PIV) . . 25 to 400 volts Average forward current (I_r)..up to 16 amps Peak surge current (one cycle)....150 amps

SPECIFICATIONS AT 25°C

Min. breakover voltage ($V_{\rm tro}$). .25 to 400 volts Max. leakage current (I_r) and (I_s).5 ma Max. forward voltage (V_r avg.). . . . 0.9 volts Max. gate voltage to fire ($V_{\rm gr}$). 3.0 volts Min. gate voltage to fire ($V_{\rm gr}$). 0.3 volts

BOOTH 2009 - IRE SHOW

RTH AMERICAN ELECTRONICS, INC.

71 Linden Street, West Lynn, Massachusetts • LYnn 8-4800

World Radio History

Product Information Service

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

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Show Hours 10 a.m. to 9 p.m. daily Monday through Thursday March 21-24, 1960

New STANDARD LINE OF RELIABILITY ENGINEERED FOR AIRCRAFT, MISSILE, AND SPACE VEHICLE APPLICATIONS

Model SIS 310242 **DC-AC Inverter** 1300VA, 115/208 Volt 30

Inputs Nom. 28VDC Outputs Nom. 115V 400cps 10 or 30 Power Ratings from 30VA to 1500VA

Space / Weight Designed To Yield Maximum Power Output **Consistent With High Reliability And Performance**

FEATURES

- PRECISION FREQUENCY
- OVERLOAD PROTECTION
- EXCELLENT WAVEFORM
- VOLTAGE REGULATED
- . PHASE LOCKED CIRCUITRY
- REVERSE VOLTAGE PROTECTION
- OUTPUT POWER RATING OUTPUT VOLTAGE SPECIAL FEATURES MODEL Precision frequency, excellent waveform, voltage regulated. \pm 1% for line, \pm 2% 30 VA 10 50 VA 10 115 VAC 400 cps SIS-40311 series ± 01 to • .05 % adjustable ± 10% SIS-40511 series load. Wide range stabilization, input 18-30 VDC, Voltage regulated $\,\pm\,1^{1}\!/_{2}\,\%$ no load to full 115 VAC 400 cps SIS-408042 series 80 VA 10 + 5 V $\pm 1\%$ load. Magnetic Amplifier voltage regulated. Rap 115 VAC 400 cps 1 % LC SIS-410042 series 100 VA 10 SIS-425041 series 250 VA 10 id on-off switching no transients high effi-+ 5% ciency. osc. tune ing fork Regulates to \pm 2% with simultaneous variation of zero to full load, and line 25 volts to 29 volts. SIS-3-425042 series SIS-3-450022 series 250 VA 30 500 VA 30 115 VAC 400 cps ± 2% ± 2% ± 1% Extreme frequency accuracy. Phase lock 750 VA 30 208/115 V 400 cps SIS-3-47512 series or 115/66.5 • .00Ż% circuitry. Magnetic voltage regulator volts Adj. ± 5% Short circuit protected, reverse voltage pro-60 VA 30 26 VAC 400 cps ± .01 % SIS-3-40613 series tection, high temp., - 100 C. Voltage reg-Adj. ± 5% ulated.

DESIGN NOTE: any of the special features described may be combined in a single unit to meet your special requirements.



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NEW

Dependability Reliability Stability

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Product Information Service

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Microwave & Radar Test Equipment

Military Equipment

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Fixed Tuned Broad Band S-Band Balanced Mixer

Operates over the entire frequency range (2.6 — 3.95 Kmc) of the WR (284) band. This new broad band balanced mixer is fixed tuned.

Freq. (Kmc)	VS at waveguide	SWR at coax	Isolation (db)	Noise figure at 0.5 ma.
2.6 2.7 2.8 2.9 3.0 3.1 3.2 3.3 3.4 3.5 3.6 3.7 3.8 3.7 3.8 3.9 3.95	2.10 1.62 1.60 1.34 1.12 1.07 1.09 1.30 1.21 1.14 1.34 1.63 1.97 2.17 2.05	1.65 1.22 1.62 2.00 1.90 1.70 1.60 1.35 1.16 1.20 1.35 1.80 2.10 1.95 1.90	18.0 21.0 24.5 26.0 28.0 27.5 29.0 21.5 26.0 21.7 22.5 23.5 23.5 23.7 21.2 21.3	6.99393034533459 5.5.55555555555555555555555555555555

Noise figure: Measured using a 1.5 db 30 mc IF strip

See us at IRE! Booth 2407



Microwave Development Laboratories, Inc. 92 Broad Street, Babson Park 57, Wellesley Mass. Telephone CEdar 5-6252 Design — Development — Production of Microwave Components and Assemblies



See us at Booth 2238

Product Information Service

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

IRE SHOW - BOOTH 1311



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Model LF-2a

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Model SSB-3a







Model TMI-la



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Numerical Listing of Exhibitors

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- 4053 GOE Engineering Co. 4054 Inso Electronic Products, Inc. 4055 Saxton Products, Inc. 4056 Chemo Products, Inc. 4101-4102 W. H. Brady Co. 4103 Alpha Wire Corp. 4104 Bodnar Industries, Inc. Inc. 4105 United Catalog Publishers, Inc. 4106 Cobehn, Inc. 4107 Electro Devices, Inc. 4108 Zell Products Corp. 4109 Diamonite Products Mfg. Co. 4110 Gries Reproducer Corp. 4111 Oryx Company 4112 Grant Pulley & Hardware Corp. 4113 Derivation & Tabulation Associates. Inc. 4114 Hull Corporation 4115 Great Eastern Metal Products Co., Div. of GEMP Manufacturing Co. 4116 International Eastern Co. 4117 Secon Metals Corp. 4118 Zero Manufacturing Co. 4119 American Sealants Company 4120 Acoustica Associates, Inc. Corp. 4121 W. L. Gore & Associates 4122 Design Tool Corp. 4123 Mitchell-Rand Mfg. Corp. 4124 Popper & Sons, Inc. 4125 International Pump & Machine Works, Inc. 4126 F. J. Stokes Corp. 4418 Haveg Industries, Inc. 4419-4421 Fairchild Publications, Inc. 4127 Standard Pressed Steel Co. 4128 McDowell Electronics, Inc. 4129 Technical Wire Products, Inc. 4130 American Aluminum Co. 4131 Robertson Electric Co., Inc. 4132 Jennings Machine Corp. 4133 Ozalid Div., General Aniline & Film Corp. 4134 Temperature Engineering Corp. 4135 Ungar Electric Tools, Inc. 4201-4203 Wilbur B. Driver Co. 4202 Boesch Mfg. Co., Inc. 4204-4206 Premier Metal Products Co. 4205 Artos Engineering Co. 4207-4211 Tinnerman Products, Inc. 4208 H. Braun Tool & Instrument Co., Inc. 4210-4212 Markem Machine Co. 4213 Hitemp Wires, Inc. 4214 W. M. Welch Manufacturing Co. 4215 Cowan Publishing Corp. 4216 Allied Chemical Corp., General Chemical Div. 4217-4219 Belden Manufacturing Co. 4218-4220 George Stevens Mfg. Co., Inc. 4221 Kester Solder Co.
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- 4322-4324 Sigmund Cohn Corp.
- 4323-4325 Leesona Corporation
- 4326 John Wiley & Sons, Inc.
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- & Co., Inc., Film Dept.
- 4330 Tensolite Insulated Wire Co., Inc.
- 4331 Green Instrument Co., Inc.
- 4401-4403 Driver-Harris Co.
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- 4416 American Super-Temperature
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- 4420-4424 Elgin Metalformers Corp.

- 4423 Birnbach Radio Co., Inc.
- 4425 The Kanthal Corp.
- 4426 Coil Winding Equipment Co. 4427 NRC Equipment Corp., National
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- 4525 Plastoid Corporation
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- 4528 Metal Fabricators Corp.
- 4529 H P L Manufacturing Co.
- 4530 Gardner-Denver Company
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- 4532 Narda Ultrasonics Corp.
- 4533-4535 Carl Hirschmann Co., Inc.

Product Information Service

The following is an alphabetical list of the basic products displayed in the show. After each heading a list of booth numbers is given, where you may see these products displayed. Use the preceding list of numerically arranged exhibitors to find the names of the companies exhibiting the products.

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March, 1960



Timing is our forte . . .

Electronic timing devices and controls that represent the state-of-the-art in performance and reliability precision performance that's built-in by engineering know-how; unquestioned reliability that's guaranteed by a comprehensive Quality Assurance Program including functional and environmental testing of each production unit. And we believe you'll appreciate the Tempo brand of customer service . . . from prompt, thorough handling of your special requirements to consistent, on-schedule deliveries and field service follow-thru. TEMPO INSTRUMENT INCORPORATED HICKSVILLE, N.Y.



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ELECTRONIC TIME DELAY RELAYS. No moving parts except relay contacts, 2PDT-2 amp or 3PDT-10 amp ratings; fixed or adjustable time delays from .02 to 300 seconds; accuracy to 3% or better; vibration-proof to 2,000 cps at 20 g's. FLASHERS. Typical operation: Flashing rate of 40 cycles per minute, with a one-to-one ratio of on time to off time.

INTERVALOMETERS. Typical operation: 300 seconds after application of 28 vdc, output relay energizes; 3 seconds later, relay de-energizes. Cycle repeats itself until supply voltage is removed. **REPEAT CYCLE TIMERS.** Provide multi-channel control signals, each with a timing pulse of pre-determined magnitude. Sequencing of output pulses is synchronized.

PULSE TRAIN GENERATORS. For control of various loads in a single system. Provides output signals and pulses of specified characteristics.

SEQUENCE TIMERS. Typical operation: Provides sixty-four sequential pulses each of 100 milliseconds duration, starting immediately upon application of 28 vdc. Automatically stops after last pulse.

PROGRAMMERS. Designed to suit particular system requirements. Typical unit may provide a complete timing and control program covering more than 5000 seconds from start.

IRE Show - Tempo Application Engineers will be at booth M8 - IRE Show.

World Radio History

Airborne Time Code Generator illustrates high-density packing obtainable with T-Series circuits.



Hinged arrangement of mounting panel facilitates accessibility.

The finished package weighs only 20 lbs.; measures 5" x 8" x 2034". Unit generates 14 digit Point Mugu code, modulating a 1 kc carrier plus a dc time code. Three sine wave and four pulse outputs are also provided, all with only 96 T-Series circuits and 77 watts of input power.



FROM SYSTEM SPECS TO BREADBOARD TO FINISHED PRODUCT IN 75 DAYS!

That's the record set by the manufacturer of this complex airborne Time Code Generator - thanks to the compatibility of proven EECO T-Series Circuit Modules and the flexibility of the EECO Breadboard Kit.

Designed and developed for testing the fire control of manned supersonic aircraft under actual flight conditions at altitudes up to 80,000 feet, this Time Code Generator employs T-Series circuits throughout. Required accuracy of 1 part in 105 was easily obtained.

HIGH DENSITY, LIGHT WEIGHT

The total package contains 96 T-Series Circuits, 14 filamenttype EECO Minisig Indicators, and power converters (the beginning of our line of compact 12-volt EECO Power supplies for use with T-Series circuits) - all within a volume of 1/2 cubic foot. In spite of this terrific packing density, the equipment still retains extreme ease of accessibility and weighs only 20 lbs. No cooling is required.

T-SERIES VS. VACUUM TUBE CIRCUITS

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WEIGHT	20 lbs. (including power converters)	160 lbs. (plus fan and power supply)		
POWER	77 watts	650 watts (plus power for fan)		

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You, too, can develop the most complex equipment in record time with these proven EECO circuits and systems development aids. They'll save you time and money in four major areas:

- 1 DESIGN You can devote full time to system design problems or unusual circuit requirements, knowing that routine circuit detail has been compatibly pre-engineered and packaged for you.
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- 3 PRODUCTION Your production problem is reduced to one of mounting sockets on panels or chassis and providing simple socket-to-socket wiring. Plug in the appropriate circuits and the system is complete.
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Why not let proven EECO T-Series circuits and systems development aids help you solve your equipment design problems?

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World Radio History

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Numerical Listing of Exhibitors

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- 3936 Republic Aviation Corp., Special Products and Services Div. 3937-3939 General Mills, Inc., Mechani-
- cal Div.
- 3938 A.R.F. Products, Inc.
- 3940 Electrol, Inc. 3943 James G. Biddle Co.
- 3944 Dynatran Electronics Corp.
- 3945 Virginia Electronics Co., Inc.
- 3946 Electronic Measurements Corp.
- 3947 Ebauches, S.A. Freeport Engineering Co.
- 3948 Zenith Radio Corp.
- 3949 Dytronics Co.
- 3950 Thermo Electron Engineering Corp.
- 3951 Unholtz Dickie Corp. 3952 Molectronics Corp., Div. Minneapolis-Moline Co.

FOURTH FLOOR

- 4001 Chassis-Trak, Inc.
 - 4002 Utica Drop Forge & Tool
- 4003 Weckesser Co.
- 4004 Barnstead Still & Sterilizer Co.
- 4005-4006 Coors Porcelain Co.
- 4007 Wright Metalcoaters, Inc.
- 4008 Edwin B. Stimpson Co., Inc. 4010 Wales-Strippit, Inc.
- 4011 MM Enclosures, Inc.
- 4012 Hexacon Electric Co.
- 4013 Cable Designs, Inc.
- 4014 Raybestos-Manhattan, Inc.
- 4015 Little Falls Alloys, Inc.
- 4016 Rosan, Inc.
- 4017 Boston Insulated Wire & Cable Co.
- 4018 Rogan Brothers, Inc.
- 4019 & Room 610 Univeral Instruments Corp.
- 4020 Dynacor, Inc.
- 4021-4022 Fansteel Metallurgical Corp.
- 4023 Allegheny Electronic Chemicals Co.
- 4024 Weller Electric Corp.
- 4025 Gudebrod Bros. Silk Co., Inc.
- 4026-4027 Monsanto Chemical Co.
- 4028-4029 Phelps Dodge Copper Products Corp., Inca Mfg. Div. 4030 Jones & Lamson Machine Co. 4031 Combined Book Exhibit, Inc.

- 4032 Gorman Machine Corp.
- 4033 American Electrical Heater Co.
- 4034 Swiss Jewel Company Herman D. Steel Co.
- 4035 New Hermes Engraving Machine Corp.
- 4036-4037 Ünistrut Products Co.
- 4038 Micromech Mfg. Corp., Div. of Sanford Mfg. Corp.
- 4039 Stewart Stamping Co.
- 4040 Adolf Meller Company
- 4041 Actioncraft Products
- 4042 Anchor Metal Co., Inc.
- 4043 Eubanks Engineering Co.
- 4044 The Zippertubing Co.
- 4045 Raytheon Co., Industrial Apparatus Div.
- 4046 Norrich Plastics Corp.
- 4047 Uniform Tubes, Inc.
- 4048 Howard W. Sams & Co., Inc.
- 4049 Optical Coating Laboratory, Inc.
 - 4050 Furane Plastics, Inc.
 - 4051 Vector Electronic Co.
 - 4052 Leach & Garner Co., and General Findings & Supply Co.

(Continued on page 468.4)

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Years of experience—for as early as 1954, General Electric had conceived and developed radar equipment capable of detecting ballistic missiles at 1,000 miles. This was the forerunner of the AN/FPS-50 surveillance radar being provided by General Electric under subcontract to RCA for the Air Force Ballistic Missile Early Warning System (BMEWS).

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DEFENSE ELECTRONICS DIVISION HEAVY MILITARY ELECTRONICS DEPARTMENT SYRACUSE, NEW YORK

World Radio History

Numerical Listing of Exhibitors

(Continued from page 462,1)

- 3123 B & F Instruments, Inc.
- 3201-3208 General Radio Co.
- 3205-3209 Polarad Electronics Corp.
- 3210-3212 F. L. Moseley Co.
- 3211 Telewave Laboratories, Inc.
- 3213 Manson Laboratories, Inc. 3214 Bogen-Presto Company, Div. of
- Siegler Corp.
- 3215-3217 Bird Electronic Corp.
- 3216-3218 De Mornay-Bonardi
- 3219-3221 Baird-Atomic 3220 Herman H. Sticht Co., Inc.
- 3222 Waveforms, Inc.
- 3223 Navigation Computer Corp.
- 3224 Wang Laboratories, Inc.
- 3225 Blonder-Tongue Laboratories, Inc. 3226-3228 Tenney Engineers, Inc. 3227-3229 Industrial Instruments, Inc.
- 3230 Muirhead Instruments, Inc.
- 3231-3233 Antlab, Inc.
- 3232 Wayne-George Corp.
- 3234 Datex Corp.
- 3235 Cubic Corp. 3236 Harvey-Wells Electronics, Inc.
- 3237 Mc Millan Laboratory, Inc.
- 3238 Julie Research Laboratories
- 3239 International Radiant Corp.
- 3240 The Geotechnical Corp.
- 3241 Telectro Industries Corp.
- 3242 John Fluke Mfg. Co., Inc. 3243 Erie Resistor Corp.
- 3244 Genisco, Incorporated
- 3301-3305 Marconi Înstruments
- 3302-3306 Hewlett-Packard Co.
- 3307-3309 Southwestern Industrial Electronics Co., Div. Dresser Inds.
- - MODEL NO. R-129 H

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HIGH POWER Ku CIRCULATOR

3308-3310 Computer Control Company,

3311 Electron Corp., Div. Ling-Altec Electronics, Inc.

3315-3317 Panoramic Radio Products,

3316-3318 Airborne Instruments Lab.,

Div. Cutler-Hammer, Inc.

3401-3405 Hewlett-Packard Co. 3402-3404 Ballantine Laboratories, Inc.

3413-3417 Airborne Instruments Lab.,

Div. Cutler-Hammer, Inc.

3416-3418 Beckman Instruments, Inc.,

3501-3503 Measurements, A McGraw-

3505 Electronic Instrument Co., Inc.

3507-3511 International Business Ma-

3515-3517 Gates Ratio Co. 3601-3605 Sanborn Co. 3602-3606 Polytechnic Research & Dev. Co., Inc., Div. Harris Intertype 3607 Millivac Instruments, Div. Cohu

Electronics, Inc. 3608-3610 Century Electronics & Instru-

3609-3611 Massa Labs., Div. Cohu Elec-

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Ínstru-

3406-3408 British Industries Corp. 3407-3409 Potter Instrument Co., Inc.

3410-3412 Sensitive Research

ment Corp.

3414 Specific Products

EICO

3411 Rixon Electronics, Inc.

Berkeley Div.

3502-3508 Collins Radio Co.

chines Corp.

3512-3518 Kay Electric Co.

3515-3517 Gates Radio Co.

ments. Inc.

tronics, Inc.

3510 Northern Radio Co., Inc.

3513 Industrial Test Equipment Co.

Edison Division

3312-3314 Crosby-Teletronics Corp.

3313 Rawson Electrical Instr. Co.

Inc.

Inc.

FREQUENCY:	15.7 to 16.9 Kmc	VSWR:	1.10 MAX.
ISOLATION:	(2% BAND) TRANSREC. 20 DB	POWER:	100 KW PEAK
INSEPTION	ANTTRANS. 20 DB	LENGTH:	3.0"
LOSS:	ANTREC. 0.5 DB	WEIGHT:	LESS THAN 12 OZ.
TEMP.:	— 40°C to + 100°C	WAVEGUIDE:	MATES WITH RG-91/U

HIGH POWER K. FERRITE SWITCH_

	See us o	it IRE Booth 3	007	KU-91/U
	I ENGTH-	+100°C 5 5″	WAVEGUIDE:	MATES WITH
MODEL NO. R-107 H	TEMP.:	- 40°C to	WEIGHT:	2 WATIS AVE. 1.0 lb.
	LOSS:	ANTREC. 0.5 DB	POWER:	70 WATTS PK.
	INSERTION	TRANSANT. 0.5 DB	DRIVING	/0 /0 SWITCHED
	ISOLATION:	TRANSREC. 20 DB ANTTRANS. 20 DB	SWITCHING TIME:	10μ sec. 90% SWITCHED
	FREQUENCY:	15.7 to 16.9 Kmc (2% RAND)	POWER:	100 KW PK.

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- 3613-3617 KinTel Division, Cohu Electronics, Inc. 3616–3618 The Hickok Electrical Instru-
- ment Co.
- 3701-3703 Gertsch Products, Inc. 3702-3706 The Technical Materiel Corp.
- 3705 Ad-Yu Electronics Lab., Inc.
- 3707-3709 Packard-Bell Electronics
- Corp., Technical Prods. Div. 3708-3710 Krohn-Hite Corp.
- 3711 MacLeod & Hanopol, Inc.
 - 3712-3718 Electronic Associates, Inc.
 - 3713-3717 FXR, Inc.
- 3801 Ferris Instrument Co.
- 3802-3804 Ling Electronics, Div. Ling-Altec Electronics, Inc.
 - 3803-3805 CGS Laboratories, Inc.
 - 3806-3808 Servo Corporation of America
 - 3807 Hastings-Raydist, Inc.
 - 3809 Precision Apparatus Co., Inc. 3810-3812 Electro-Pulse, Inc.

 - 3811 Systron Corporation
 - 3813 The Narda Microwave Corp., HPED Division
 - 3814-3816 Burlingame Associates
 - 3815-3817 The Narda Microwave Corp.
 - 3818-3820 Empire Devices Products Corp.
 - 3819-3821 Laboratory for Electronics, Inc.
 - 3822-3824 Bell Aircraft Corp., Avionics Division
 - 3823 Beattie-Coleman, Inc.
 - 3825 Victor Adding Machine Co.
 - 3826-3828 Nems-Clarke Co., Div. Vitro Corp. of America 3827-3829 Wayne Kerr Corporation

 - 3830-3832 Associated Testing Labs., Inc.
 - 3831 Digital Equipment Corp.
 - 3833 North Atlantic Industries, Inc.
 - 3834 Rutherford Electronics Co.
 - 3835 Franklin Electronics, Inc.
 - 3836-3838 Electro Instruments, Inc.
 - 3837-3839 Computer Systems, Inc.
 - 3840-3842 Acton Laboratories, Inc. 3841 The United States Time Corp.

 - 3843 Menlo Park Engineering 3843B Frequency Standards, Inc., Div. National Electric Products Corp. 3844 Hoover Electronics Co., Subsid.
 - The Hoover Co. 3846 Micro Gee Products, Inc.

 - 3848 Conrad, Inc. 3901-3905 Allen B. DuMont Labs., Inc.
 - 3904-3906 Epsco, Inc.

 - 3907 Keithley Instruments, Inc.
 - 3908 Veeder-Root, Inc. 3909 Scientific-Atlanta, Inc.

 - 3910-3914 American Bosch Arma Corp. 3911-3913 Lavoie Laboratories, Inc.
 - 3914B Edgerton, Germeshausen & Grier, Inc.
 - 3915-3917 Knight Electronics

Mincom Div. 3924 Greibach Instruments Corp. 3925-3927 American Machine & Foundry

Co. 3926 The Garrett Corp.

Inc.

3934 Centronix, Inc.

Inc.

- 3916-3918 Assembly Products, Inc.
- Metronix Div., Assembly Prods.
- 3919-3921 The Martin Co. 3920-3922 California Technical Industries, Div. Textron, Inc. 3923 Minnesota Mining & Mfg. Co.,

3928-3930 Ace Engineering & Machine

Solar Sol

3932 Page Communications Engineers,

Inc., Div. Northrop Corp.

3935 American Electronic Laboratories,

(Continued on page 466.4)

March, 1960

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Astron's new 50 volt hermetically sealed subminiature paper capacitors have the reliability required by specification MIL-C-25A. These units operate at temperatures from -60°C to +125°C without derating. The capacitance variation is less than $\pm 3\%$ over the entire operating temperature range. High insulation resistance, low power factor, unusually low resonance loss are com-bined in this new light-weight, subminiature unit.

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Numerical Listing of Exhibitors

(Continued from page 16021)

THIRD FLOOR

- 3000 Rohde & Schwarz Sales Co. (U.S.A.) Inc. 3001-3002 Veeco Vacuum Corporation
- 3003 Levinthal Electronic Prods. Inc.
- 3004 Radiation, Inc.
- 3005 Radar Measurements Corp.
- 3006 Wallson Associates, Inc.
- 3007 Strand Labs., Inc.
- 3008 Blue M Electric Company
- 3009 Mid-Eastern Electronics, Inc. 3010-3011 Electro-Measurements, Inc.
- 3012 Landis & Gyr, Inc
- Sodeco
- 3013 Foto-Video Laboratories, Inc.
- 3014 Instron Engineering Co.
- 3015 Sierra Research Corporation
- 3016 Optimized Devices, Inc.
- 3017 Magnetic Instruments Co., Inc., Sub. Pyrometer Co America. Inc.
- 3018 ELIN Division, International Electronic Research Corp.
- 3019-3020 Dymec, Div. of Hewlett-Packard Co. 3021 The Standard Electric Time Corp.
- 3022 Telonic Industries, Inc.
- 3023 North Hills Electric Co., Inc.
- 3024 Pitometer Log Corp. 3025 EH Research Laboratories, Inc.
- 3026 General Resistance, Inc.
- 3027-3030 Tektronix, Inc. 3031-3032 Sierra Electronic Corp.

- 3033 Trio Laboratories, Inc.
- 3034 Quan-Tech Laboratories
- 3035 Precision Instrument Company 3036 Barnes Engineering Co.
- 3037 Precise Development Corp.

- 3038-3039 Hermes Electronic Co. 3041-3042 Non-Linear Systems, Inc. 3043 Temco Aircraft Corp., Electronics Div.
- 3044 Pickard & Burns, Inc. 3045 Di/An Controls, Inc.
- 3051 Offner Electronics, Inc.
- 3052 Burr-Brown Research Corp.
- 3053 Itemco, Inc.
- 3054 Yokogawa Electric Works, Inc.
- 3056 Jerrold Electronics Corp.
- 3059 Lumatron Electronics, Inc.
- 3060 General Dynamics Corp., Liquid Carbonic Div.
- 3061-3062 Shielding, Inc. 3063 CBS Laboratories, Div. Columbia Broadcasting System, Inc.
- 3064 Avco Corporation, Crosley Division
- 3065 Avco Corporation, Research &
- Adv. Development Div. 3101-3102 Boonton Radio Corp.
- 3103-3104 Computer-Measurements
- Corp. 3105-3106 Waterman Products Co., Inc.
- 3107-3109 MB Electronics
- 3108 Northeastern Engineering, Inc.
- 3110 Cook Electric Co., Data-Stor Div.
- 3111 Philamon Labs., Inc. 3112-3113 Electronic Tube Corp.
- 3114 Boonton Electronics Corp.
- 3115-3119 Radio Frequency Laboratories, Inc.
- 3116-3118 Probescope Company
- 3120 Dynapar Corporation
- 3121 Larson Instrument Co.
- 3122 Precision Scientific Co.

(Continued on page 461.1)

MICROWAVE CAVITIES

Here is a cavity, designed for microwave link application, which replaces a unit costing four times as much. Shown are glass seal tunable cavities for use with microwave discriminators. Some of their outstanding characteristics are:

- 1. Tuning range 5925-6425 mc 6575-6875 mc 7125-7750 mc
- 2. Temperature stability ± 2 mc for - 20° to +70°C.
- 3. No change with humidity.
- 4. Can withstand severe vibration and shock.
- 5. Loaded Q = 250.

This unit along with its accompanying discriminator, can be made available in all waveguide sizes from L through Ka band.



HOW DO YOU TEST AN X-Y PLOTTER

In an automobile, transmissions can be designed to give top speeds by sacrificing acceleration—but it's a poor bargain, as a quick road test will show. Similarly, in an X-Y analog plotter high slewing speeds can be obtained by sacrificing acceleration. Again it's a poor bargain because highest plotting accuracy depends upon high static accuracy combined with a *perfect balance between acceleration and velocity limits*. EAI's Model 1100E Variplotter has this desired balance as a simple 'road test' developed by EAI engineers can graphically demonstrate.

As a matter of fact, everything about the Model 1100E Variplotter has been engineered to give you the utmost in plotting performance. Developed to speed up engineering control, testing and design operations, it consistently produces faster, more accurate plots of X-Y related data.

The development of the Model 1100E has resulted from EAI's years of pioneering research and development in the field of automatic plotting. It provides outstanding accuracy of 0.075% F.S. – less than the width of the line drawn by the pen. Arm acceleration



is 250 inches/sec.². Pen acceleration is 750 inches/sec.². The high velocity of the 1100E is augmented by this faster acceleration to assure outstanding dynamic performance.

Repeated testing under actual operating conditions proves that the principle of the Variplotter, Model 1100E design virtually eliminates backlash, and provides drift-free operation for periods of 8 hours or more. This superior repeatability has been amply testified by users who report that even after overnight shut-down, the Model 1100E resumes plotting with no noticeable drift.

A complete line of accessories—including bi-variant function generator, digital data plotting (manual or automatic) and time base generator—makes the EAI Variplotter the most versatile automatic plotting method available. The Model 1100E can be easily converted to operate as a function generator—or will plot digital information manually from a keyboard as well as automatically from punched cards or paper tape—by simple addition of compatible components.

Check these features...

Portable desk-top size – Large plotting surface (11" x 17") – Vacuum hold down – High dynamic and static accuracy – Rugged construction – Ease of maintenance – Differential input – Plug-in input network – Superior repeatability.

Ask your EAI representative to show you the simple laboratory test that proves the superiority of the Model 1100E Variplotter,—or write for Bulletin AP 810-1. See Model 1100E Variplotter in operation, N.Y. IRE. Booths 3712-18.

ELECTRONIC ASSOCIATES, INC. Long Branch, New Jersey

Numerical Listing of **Exhibitors**

(Continued from page 458A)

- 2305 Tru-Ohm Products Div. Model Engineering & Mfg., Inc. 2306-2308 Airpax Electronics Incorpo-
- rated 2307-2309 Continental Connector Corp. De Jur-Amsco Corp.
- 2310 Illinois Condenser Co.
- 2311 Mc Coy Electronics Co. 2312 Electro-Mec Laboratory, Inc.
- 2313 Curtis Development & Mfg. Co.
- 2314-2316 Allen-Bradley Co.
- 2315 Microtran Co., Inc.
- 2317-2319 Technology Instrument Corp.
- 2318-2320 Lambda Electronics Corp.
- 2321-2323 American Television & Radio
- Cu. 2322-2332 Sylvania Electric Products,
- Inc. 2325 Herman H. Smith, Inc.

- 2325 Herman H. Smith, Inc.
 2327 Stevens Mfg. Co., Inc.
 2329-2331 M. C. Jones Electronic Co., Inc., Sub. Bendix Aviation Corp.
 2333-2335 Ohmite Manufacturing Co.
 2334-2336 Corning Glass Works
 2337 Donald P. Mossman, Inc.

- 2338 Federal Pacific Elec. Co., Fifty Avenue L Div.

- Avenue L Div. 2339 Bogart Manufacturing Corp. 2340 Phillips Control Corp. 2341 Buchanan Electrical Prods. Corp. 2342-2344 John Gombos Co., Inc. 2343 A'G'A Division, Elastic Stop Nut Corp. of America
- 2401-2405 Union Carbide Consumer Products Co., Div. of Union Carbide Corp.
 - Kemet Co., Div. Union Carbide Corp.
- 2402-2408 Amphenol-Borg Electronics Corp
 - Borg Equip. Div.
 - Industrial Prods-Danbury Knudsen

 - Amphenol Connector Corp. Amphenol Cable & Wire Div.
 - Amphenol Pacific Div.
- 2407 Microwave Development Labs., Inc. 2409 Price Electric Corp.
- 2410-2412 Eitel-McCullough, Inc.
- 2413-2414 United Transformer Corp.
- 2415-2425 Sylvania Electric Products.
- Inc.
- 2416-2424 Sprague Electric Co. 2426 The Triplett Electrical Instrument Co.

- 2427-2429 Corning Glass Works 2428-2430 Tung-Sol Electric, Inc. 2431 Balco Research Laboratories, Inc.
- 2432 Sperry Electronic Sperry Rand Corp. Div., Tube
- 2433 Dage Electric Co., Inc.
- 2434 Sperry Semiconductor Div., Sperry Rand Corp.
- 2435 Lieco, Inc.

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- 2436 Sperry Gyroscope Co., Div. of Sperry Rand Corp.
 2437 Magnetics, Inc.
- 2438 Sperry Microwave Electronics Co., Div. Sperry Rand Corp.
- 2501-2507 Amphenol-Borg Electronics Corp.
- 2502-2504 Guardian Electric Mfg. Co.
- 2506-2508 Robinson Technical Products, Inc.
- 2509-2515 The Arnold Engineering Co.
 - Be sure to see all four floors!

- 2510-2520 International Telephone & Telegraph Corp. Federal Electric Corp. ITT Components Div. ITT Federal Div. ITT Industrial Prods. Div. ITT Laboratories International Electric Corp. Kellogg Switchboard & Supply Co. 2517 James Millen Mfg. Co., Inc. 2519 Audio Devices, Inc. 2521-2523 Tung-Sol Electric, Inc. 2522-2524 Amperex Electronic Corp. 2525 Magnecraft Electric Co. 2526-2528 Electrical Industries, Div. Philips Electronics, Inc. 2527-2531 AMP Incorporated Capitron Div., AMP, Inc. 2530 Ferroxcube Corp. of America 2532 D. S. Kennedy & Co. 2533 Magnetics, Inc. 2534 Barry Controls Incorporated 2535-2536 Cinch Mfg. Corp. Howard B. Jones Div., Cinch Mfg. Corp. United-Carr Fastener Corp. 2601 Air-Marine Motors, Inc. 2602 Astron Corp. 2603-2607 Aerovox Corporation 2604-2614 Raytheon Company Machlett Laboratories, Inc. Sorensen & Co., Inc. Telephone & 2615-2625 International Telegraph Corp. 2616–2626 Clevite Corp. Brush Insts. Div. Clevite Electronic Components Clevite Transistor Products 2627-2629 Dale Products, Inc. 2628-2632 Cannon Electric Company 2631 Gorham Electronics Div., Gorham Mfg. Co. 2633 Globe Industries, Inc. 2634 Shallcross Mfg. Co. 2635 San Fernando Electric Mfg. Co. 2636-2638 Kepco, Inc. 2637 D. S. Kennedy & Co. 2701-2707 Fairchild Camera & Instrument Corp. Fairchild Controls Corp. Fairchild Semiconductor Corp. 2702-2704 Potter & Brumfield, Inc., Div. American Machine Foundry 2706 Belling & Lee Limited Ercona Corp. 2708 The James Knights Co. 2709 Behlman Engineering Co. 2710–2712 Bomac Laboratories, Div. Varian Associates 2711 Syntronic Instruments, Inc. 2713-2715 U.S. Semiconductor Products, Div. United Industrial Corp. 2714–2720 Varian Associates 2717-2719 The Daven Company 2721-2723 Freed Transformer Co., Inc. 2722-2732 The Superior Electric Co. 2725-2727 Cornell-Dubilier Electric Corp., Affil. Federal Pacific Electric Co. 2729-2731 Mycalex Corp. of America 2733-2735 Sigma Instruments, Inc. 2734 The Electro Motive Mfg. Co., Inc. 2736 The Capitol Machine Co.
- 2737 Bussmann Mfg. Division, McGraw-Edison Co. 2738 RS Electronics Corp.
- 2739-2741 Thomas A. Edison Industries, Instrument Div.
- 2740 Plastic Capacitors, Inc.
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- 2803 Penta Laboratories, Inc.
- 2805 U.S. Components, Inc.
- 2806 Burgess Battery Co.
- 2807 Mucon Corp.

- 2808 Filtors, Inc. 2809-2811 Communication
- Accessories Co., Subsidiary of Collins Radio Co
- 2810 Hardwick, Hindle, Inc.
- 2812 Servomechanisms, Inc., Mechatrol Div.
- 2813 International Instruments, Inc.
- 2814-2816 Garlock Electronic Products, The Garlock Packing Co.
- 2815-2819 Sola Electric Co., Div. Basic Products Corp.
- 2818 ERA Electric Corp.
- 2820 Electronic Research Assoc., Inc.
- 2821-2823 Kings Electronics Co., Inc.
- 2822-2826 International Resistance Co.
- 2825 Kemtron Electron Products, Inc.
- 2827 Switchcraft, Inc.
- Corp., 2828 Perkin-Elmer Vernistat Div.
- 2829-2831 Dialight Corp.
- 2830-2832 Rotron Mfg. Co.
- 2833 J-B-T Instruments, Inc.
- 2834-2836 Electra Manufacturing Co.
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- 2837 Industrial Electronic Hardware
- Corp. 2838 The Gamewell Co., Potentiometer Div.
- 2839 Polyphase Instrument Co.
- 2840 Bergen Laboratories 2841–2843 Heinemann Electric Co.
- 2842 Bradley Semiconductor Corp.
- 2844 Northeast Scientific Corp.
- 2900 Kulka Electric Corp.
- 2901-2903 International Rectifier Corp.

2904 General Electric Co., Semiconduc-

2906-2908 General Electric Co., Receiv-

2912-2914 General Electric Co., Power

2913 Thomas Electronics, Inc. 2914 General Electric Co., Laminated

2916 General Electric Co., Large Lamp

2920 General Electric Co., Missile and

2924 General Electric Co., Heavy Mili-tary Electronics Dept., Voltage

2925-2926 Rheem Semiconductor Corp.

2928 General Electric Co., Capacitor

2930-2931 The Carborundum Co. 2932 General Electric Co., Industrial

2933 Richard Hirschmann Radio Tec-nisches Werk, Rye Sound Corp.

2935 Dempa Shinbun, Inc. Japan Electric, Toko Coil Labs., Tokyo Shibaura Electric Co., Ltd.

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March, 1960

Space Vehicle Dept. General Electric Co., Magnetic

General Electric Co., Power Trans-

General Electric Co., Specialty

General Electric Co., Instrument

Heating Dept. General Electric Co., Silicone

2917-2918 Huggins Laboratories, Inc.

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2937 Shepherd Industries, Inc.

2934 Stevens-Arnold, Inc.

2927 Comar Electric Co.

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2921 Oak Manufacturing Co.

2904-2932 General Electric Co.

tor Products Dept. 2905-2907 Allied Control Co., Inc.

ing Tube Dept. 2909–2910 Burnell & Co., Inc.

2911 Christie Electric Corp.

Products Dept.

Tube Dept.

2915 ESC Corporation

Dept.

2922 Sealectro Corp.

2923 Littelfuse, Inc.

2919 Telex, Inc.



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Take other areas of special interest to Stromberg-Carlson engineers: Instrumentation and safety systems for nuclear reactors and ground testing equipment for missile systems. Here interchange of information with General Atomics, Electric Boat and Convair Divisions adds a new dimension to Stromberg-Carlson's electronics capability.

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Also positions for:

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STROMBERG-CARLSON

1476 N. Goodman St., Rochester 3, New York

World Radio History

Numerical Listing of Exhibitors

(Continued from page 457A)

- M-6 Renbrandt, Inc.
- M-7 Faradyne Electronics Corp., Affil. Mansol Ceramics
- M-8 **Tempo Instrument Incorporated**
- M-9 Antenna & Radome Research
- Assoc. M-10 Syntron Company
- M-11 Universal Electronics Co.
- M-12 Pesco Products Division **Borg-Warner Corporation**
- M-13 Nothelfer Winding Laboratories, Inc.
- M-14 Tru-Connector Corp.
- M-15 Babcock Relays, Inc

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M-16 Master Specialties Co.

- M-17 Hoyt Electrical Instrument Works, Inc.
- M-18 Chicago Dynamic Industries, Inc., **Precision Products Division**
- M-19 Lindberg Engineering Co., High Frequency Division
- M-20 Ecco Electronic Corp.
- M-21 Pittman Electrical Developments Co.
- M-22 Security Devices Lab., Div. Sargent & Greenleaf, Inc.
- M-23 Chicago Aerial Industries, Inc., **Kintronic** Division
- M-24 D & R Ltd.
- M-25 Granite State Machine Co., Inc.

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- 2004 Sony Corporation
- 2005 Moeller Instrument Company, Inc., Electronics Div.
- 2006 Murata Mfg. Co. Ltd., International Div.
- 2007 National Semiconductor Corp. 2008 Microwave Electronic Tube Co., Inc.
- 2009 North American Electronics, Inc.
- 2100 James Electronics, Inc.
- 2101-2103 Microdot, Inc.
- 2102 LEL, Inc.
 - 2104 Power Designs, Inc.
 - 2105 G-M Laboratories, Inc.
 - 2106 Instrument Development Labs., Inc.
 - 2108 Analogue Controls, Inc.
 - 2109 Carter Parts Co.
- 2110-2118 Engelhard Industries, Inc.
- 2113 La Pointe Industries, Inc.
- 2115-2117 Bogue Electric Mfg. Co.
- 2119 George Ulanet Co.

- 2120 Tech Laboratories, Inc. 2121-2123 Winchester Electronics, Inc. 2122-2124 Union Switch & Signal Div.,
- Westinghouse Air Brake Co.
- 2125 North Electric Co.
- 2126 G-C Electronics Mfg. Co.
- 2127 Yardney Electric Corp.
- 2128 Telecomputing Corp. 2129 The Gudeman Co.
- 2130 Electralab Printed Electionics Corp.
- 2130A Rowan Controller Co.
- 2131 Lord Manufacturing Co.
- 2132-2133 Eastern Industries, Inc.
- 2134 Kurman Electric Co., Sub Crescent Petroleum Corp. Subs. of
- 2201-2203 Photocircuits Corp.
- 2202-2214 Minneapolis-Honeywell Regulator Co.
- U.S. Stoneware Co., 2205-2207 The Alite Div.
- 2209 Vari-L Company, Inc.
- 2211 Drake Manufacturing Co.
- 2213-2215 Electronic Measurements Co., Inc.
- 2216 Southern Electronics Corporation
- 2217 Gremar Manufacturing Co., Inc. 2218-2220 C. P. Clare & Co.
- 2219 Cambridge Thermionic Corp.
- 2221-2225 Simpson Electric Company
- 2222-2232 Bendix Aviation Corp.
 - Scintilla Division
 - Red Bank Div
 - Bendix-Pacific Div.
 - Montrose Div.
 - Eclipse-Pioneer Division
- Semiconductor Products
- 2227 Hi-G, Inc. 2229 IMC Magnetics Corp.
- 2231 Ward Leonard Electric Co. 2233 Elgin National Watch Co., Advance Relays Division
- 2234 The Victoreen Instrument Co.
- 2235 Pyrofilm Resistor Co., Inc.
- 2236 Sage Electronics Corp.
- 2237 James Cunningham Son & Co., Inc.
- 2238 The Ericsson Corp.
- 2239 Epco Products, Inc.
- 2240 Condenser Products Div., New
- Haven Clock and Watch Co.
- 2241-2243 Douglas Microwave Co., Inc. 2242-2244 Barber-Colman Co.

(Continued on page 460A)

March, 1960

- 2301-2303 Microwave Associates, Inc.
- 2302-2304 Vitramon, Inc.

Numerical Listing of Exhibitors

(Continued from page 456A)

- 1728B Consolidated Diesel Electric
- Corp. 1729 The Thomas & Betts Co., Inc.
- 1730 Sarkes Tarzian, Inc.
- 1731 Varo Mfg. Co., Inc.
- 1801-1803 Heath Co., Div. of Daystrom,
- Inc. 1802-1804 Jennings Radio Mfg. Corp.
- 1805 Daystrom Transcoil, Div. of Day-
- strom, Inc.
- 1806-1808 Industrial Timer Corporation 1807-1809 Daystrom Inc.
- Weston Instruments 1810 Budd-Stanley Co., Inc.
- 1811-1813 Ace Electronics Associates, Inc.
- 1812-1814 Filtron Co., Inc.
- 1815 Mitronics, Inc.
- 1816 Del Electronics Corp. 1817 The Bristol Co.
- 1818 Electronic Specialty Co., Technicraft Division
- 1819-1823 The Singer Manufacturing Co., Military Products Division HRB-Singer, Inc.
- 1820 British Radio Electronics Ltd.
- 1822 The Walkirt Co.
- 1824 Twin Lock, Incorporated
- 1825 Harrison Laboratories, Inc.
- 1900 Telemeter Magnetics, Inc. 1901 Sonotone Corp.
- 1902 Control Electronics Co., Inc.
- 1903 Hill Electronics, Inc. 1904 General Magnetic Corp.
- 1905 Filmohm Corp.
- 1906-1908 Automatic Sales Electric Corp.
- 1907-1909 Spectrol Electronics Corp. 1910 Sage Laboratories, Inc.
- 1911 Keystone Products Co.
- 1912 National Cash Register Co., Electronics Div.
- 1913 Minnesota Mining & Mfg. Co. Magnetic Products Div.
- 1914 Power Sources, Inc.
- 1915 Vemaline Products Co.
- 1916 G-L Electronics Co., Inc.
- 1917 Dyna-Empire, Inc.
- 1918 Aladdin Electronics Div.
- Aladdin Industries, Inc.
- 1919 The Torrington Manufacturing Co. 1920-1924 Hoffman Electronics Corp.,
- Semiconductor Div. 1921 Sangamo Electric Co.
- 1923 Branson Corp.
- 1925 Walter Kidde & Company, Inc.
- 1926 Reon Resistor Corp. 1927 American Electronics, Inc.
- 1928 Columbus Electronics Corp.
- 1929 Electro Tec Corp.
- 1930 Deluxe Coils, Inc.

FIRST MEZZANINE

- Aetna Electronics Corporation M-1
- Guidance Controls Corp. M-2
- Methode Manufacturing Corp. M-3
- The Pyle-National Co. M-4 M-5 Slip Ring Co. of America

(Continued on page 458A)

First Aid Room

A nurse is in charge at all times. First aid room is located on the first floor mezzanine, northwest corner of the first floor. Take elevator 20.

RESEARCH AND DEVELOPMENT ENGINEERS

Interested in - DIRECT CONVERSION OF HEAT INTO ELECTRICITY

For the past several years THERMO ELECTRON ENGINEERING CORPORA-TION has been engaged in the research and development of thermionic converters and magnetohydrodynamic systems. Because of greatly increased activity on these advanced projects, we are pleased to announce the following immediate openings:

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General ability in light mechanical devices. Able to design jigs, fixtures and mechanisms for research work. At least five years experience. Broad knowledge of vacuum tube technology highly desirable. Work to be accomplished in the research laboratory with technical personnel directly involved in experiments.

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To direct and coordinate research projects. Broad experience in vacuum tube technology necessary. Must have administrative or supervisory experience. Experience with metal or ceramic vacuum tubes highly desirable.

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To design advanced systems utilizing direct conversion schemes. Must have analytical experience in heat transfer and strength of materials. Experience in high temperature applications desirable. Work includes preliminary design through to final prototype analysis.

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To design and build small, high temperature combustion chambers and burner mechanisms to operate under extreme environmental conditions of ambient temperature. Five years experience and broad technical background in combustion desired.

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To design, execute and interpret experiments in vacuum tubes. Ability to design and build electronic circuits for measuring very small currents necessary. Broad experience in vacuum tube technology and preference for experimental work highly desirable.

PHYSICS MAJOR, M.S. LEVEL OR EQUIVALENT

To perform theoretical analysis and associated experiments in electron and ion dynamics. Broad experience in field of vacuum technology and familiarity with metallurgy and materials in vacuum tube technology highly desirable.

In each case, salary is commensurate with education and experience. To arrange for an interview, please submit a resume to Lawrence T. Sullivan, Business Manager at the address indicated below:



professional opportunities at Honeywell Aero APPLIED RESEARCH

This department must expand 100% in the next three years to explore many new areas and provide direction for the fast growing Aeronautical division. Opportunities for engineers and scientists exist in these areas of current investigation: computer systems, op-tics, inertial sensors, human factors, systems analysis, instrumentation and automatic controls. A few of the specific requirements are:

INSTRUMENTATION ENGINEERS:

engineering physicists to investigate new instrumentation concepts, conduct experiments and make comparative evaluations of instrumentation feasibility.

SYSTEMS ANALYST:

capable of conducting research studies involving new techniques of space navigation and guidance.

ASTRO PHYSICIST:

for analysis of physical phenomena in space flight, including energy absorption and conversion studies.

DIGITAL EQUIPMENT ENGINEER: for research in digital logic or circuitry for application in naviga-tion and guidance systems.

ELECTRON DEVICE PHYSICIST:

capable of independent research in molecular physics connected with generation of radiation and/or plasma devices.

PROGRAMMER ANALYST:

mathematician with experience in the use of medium and large scale digital computers for analysis of scientific problems.

HUMAN FACTORS ENGINEER:

capable of analysis and direction of experiments in human motor skills, and application to manmachine systems involving automatic control techniques.

If you desire to investigate any of the above professional opportuni-ties at the Aeronautical Division, please write in confidence to Hugo Schuck, Director of Aero Research, Dept. 470C.



AERONAUTICAL DIVISION

2600 Ridgway Raad, Minneapalis 13, Minnesata

To explore professional opportunities in other Honeywell operations coast to coast, send your application in confidence to H.K. Eckstrom, Honeywell, Minneapolis 8, Minn.

Numerical Listing of Exhibitors

(Continued from page 454A)

- 1501-1503 Librascope, Incorporated
- 1502-1504 Andrew Corporation
- 1505 GPL Div., General Precision, Inc.
- 1506 Licon Switch & Control Division Illinois Tool Works
- 1507 Link Aviation
- Div. of General Precision Inc.
- 1508-1510 Alden Products Co.
- 1509-1511 Kearfott Company, Inc. 1512 Alden Elect. & Impulse Recording Eq. Co.
- 1513 Bowmar Instrument Corp.
- 1514 Alfax Paper and Engineering Co.
- 1515 Elco Corp.
- 1516 Phaostron Instrument & Electronic Co.
- 1517 Glasseal Products Co., Inc.
- 1518 Maurey Instrument Corp.
- 1519-1519A Curtiss-Wright Corp., Electronics Div.
- 1519B Solid State Products, Inc.
- 1520 E.M.I.-Cossor Electronics Ltd.
- 1521-1525 Thompson Ramo Wooldridge Inc
- 1522 International Electronic Research Corp.
- 1524 Automatic Metal Products Corp.

- 1600 Stepper Motors Corp. 1601–1607 Westinghouse Electric Corp. 1602–1608 Radio Corporation of America Electron Tube Div.
- 1609-1615 Hughes Aircraft Co.
- 1610-1618 Litton Industries, Inc.
- 1617 Tower Construction Co.
- 1619-1621 Bourns, Inc.
- 1620 M. Ten Bosch, Inc.
- 1622 JFD Electronics Corp.
- 1623 Audio Development Co.
- 1624 McLean Engineering Labs.
- 1625 Pic Design Corp. 1626 Ortho Filter Corp.
- 1627 C & K Components Inc.
- 1627A The Hart Manufacturing Co.
- 1627B United Mineral & Chemical Corp.
- 1628 Industro Transistor Corp.
- 1629-1631 Thompson Ramo Wooldridge. Inc.
- 1630 Sinclair Radio Labs. Limited
- 1632 Electro-Voice, Inc.
- 1633 Alfred Electronics
- 1701-1707 Radio Corporation of America Electron Tube Div.
- 1702 Heath Co.
- Div. of Daystrom, Inc. 1704-1706 Daystrom Pacific Div. of Daystrom, Inc.
- 1708-1710 Daystrom Inc.
- Weston Instruments
- 1709-1717 Litton Industries, Inc. 1712-1714 Wind Turbine Co.
- 1716-1718 Alford Manufacturing Co., Inc.
- 1719-1721 Bulova Res. & Dev. Labs., Inc.
- Bulova Watch Co. 1720-1722 Gabriel Electronics Div., The Gabriel Co.
- 1723 Sanders Associates Inc.
- 1724 Ceramaseal, Inc.
- 1725 Deltime, Inc.
- 1726A General Time Corp.
- 1726B Schutter Microwave Corp.
- 1727 Hetherington Switch Div., Controls Co. of America
- 1727A-1727B Electrosnap Switch Div., Controls Co. of America 1728A Consolidated Avionics Corp.

(Continued on page 457A)

- OPPORTUNITY
 - to be a part of a growing company
 - FACILITIES and equipment to carry out the job
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 - COMPENSATION for your achievements
 - SECURITY with a progressive well established company

ELECTRONICS PROJECT ENGINEER

Experience should include circuit design, video techniques, r-f design, pulse techniques and similar activities.

PHOTOSURFACE PHYSICIST

Well grounded in development, formation and measurement of photosensitive surfaces. Back-ground should include high vacuum technique, tube design, electron optics and associated skills.

Please send resume to:

General Manager

Diamond Electronics Div. of Diamond Power Specialty Corp. Box 415, Lancaster, Ohio



see full-page ad page 395A



NEW IDEAS IN RADIO-ELECTRONICS

Year after year, the IRE NATIONAL CONVEN-TION AND RADIO ENGINEERING SHOW gets bigger! That's because you and your gigantic radio-electronics industry are surging ahead with NEW IDEAS and remarkable speed to make the Space Age the most exciting time in which to live.

That's why it takes all *4 floors* of New York's great Coliseum to show what your industry is doing. Takes 950 exhibitors...takes over 200 papers... takes over 60,000 of your co-workers to view the impressive sight.

If you're not at the IRE CONVENTION AND SHOW this year you'll miss a once-a-year opportunity unequalled in your industry to see progress in action. Plan to be at the Coliseum to see...to hear about...the NEW IDEAS IN RADIO-ELECTRONICS, 1960!

The IRE NATIONAL CONVENTION Waldorf-Astoria Hotel and The RADIO ENGINEERING SHOW Coliseum, New York City



The Institute of Radio Engineers 1 East 79th St., New York 21, N. Y. 00

PULSE DOPPLER **RADAR ENGINEER**

Applicant must have a BS/EE degree plus graduate work in mathematics and electrical networks. Should have minimum of three years'experience in the design of radio frequency "front ends" for pulse doppler radars, with specific development experience in at least two of the following:

- Highly stable, high frequency oscillators.
 High power, fast recovery duplexers.
- 3) Master oscillator - power amplifier (MOPA) tube chains at microwave frequencies.

Familiarity with high power travelling wave tubes, backward wave amplifiers and backward wave oscillators is desirable.

Successful applicant will be responsible for generation of system specifications, initial design, and development of all facets of radio frequency circuitry for pulse doppler radars.



or communications engineering, with emphasis on electromagnetic theory and advanced mathematics. He must also have at least two years' experience beyond school in fundamental design and formal mathematical analysis of microwave antennas. Of particular importance is design experience with flush mounted antennas, two dimensional arrays, electronically scanned antennas, and dielectric lenses. Experience with monopulse antennas and microwave phase shifting techniques is also desirable.

The engineer who qualifies for this position will be responsible for leading initial antenna research and design, and for conceiving satisfactory theoretical solutions to radiation and scanning problems with only fragmentary information.

These career opportunities are immediately available at Emerson Electric's modern facilities in suburban St. Louis, Missouri.

In a climate of creative freedom, continual expansion and clearly outlined programs for the future, our engineers are daily influencing the state of the art. We emphasize research, design, and development with a healthy balance of production - both military and commercial.

Emerson Electric is a well-established dynamic organi-zation with 900 engineers and 5,000 employees. Salaries and benefits are top level. Our beautiful suburban location is ideal in every way. All moving expenses are fully paid.

If either of the above positions holds genuine appeal, wouldn't it pay to investigate the possibilities for you at Emerson?

Write in strict confidence to A. L. Depke.



Numerical Listing of Exhibitors

(Continued from page 452.4)

1217 Airflyte Electronics Co. 1218-1224 General Instrument Corp. Automatic Manufacturing Division Defense Products Div. Semiconductor Division Micamold Electronics Mfg. Corp. Radio Receptor Co., Inc. F. W. Sickles Company 1219 Engineered Electronics Co. 1221 Ainslie Corp. 1223 Hermetic Seal Corp. 1225-1227 Standard Electrical Products Co. 1226 Delco Radio Division General Motors Corp. 1228 Sterling Precision Corp. 1229 A.P.M. Corporation 1230-1232 Travco Associates 1231 Electro-Mechanical Instrument Co. 1233 Waters Manufacturing, Inc. 1234 Rotating Components, Inc. 1301-1309 Dynamics Corporation of America 1301-1303 Radio Engineering Laboratories, Inc. 1302-1308 Philco Corp. Lansdale Tube Co. Div. 1305-1307 Reeves Instrument Corp. 1309 Reeves-Hoffman Div., Dynamics Corp. of America 1310-1312 General Ceramics Div. Indiana General Corp. 1311 Pyramid Electric Company 1313-1315 C W S Waveguide Corp. 1314-1316 Indiana Steel Products Div. Indiana General Corp. 1317 Telrex Labs. 1318 Bliley Electric Company 1319–1323 Transitron Electronic Corp. 1320-1322 L. L. Constantin & Co. 1324 The Korfund Co., Inc. Federal Div. 1325 Craig Systems, Inc.1326 Arthur Ansley Manufacturing Co.1327 Crucible Steel Co. of America 1327A Ind. Develop. Eng. Assoc. Inc. 1327B Reeves Soundcraft Corporation 1328 National Union Electric Corporation 1329-1331 Burndy Corporation Omaton Div. 1330-1332 John Oster Mfg. Co. Avionic Division 1333 PCA Electronics Incorporated 1400 Chicago Telephone Supply Corp. 1401-1407 National Co., Inc. Servo Dynamics Corp. 1402-1408 Westinghouse Electric Corp. 1409-1421 Texas Instruments Incorporated Apparatus Div. **Component Sales** 1410-1412 P. R. Mallory & Co., Inc. 1414 Radio Materials Co. Div. of P. R. Mallory & Co., Inc. 1416–1418 Perkin Engineering Corp. 1420–1422 The A. W. Haydon Co. 1423 Victory Engineering Corp. 1424 Instruments for Industry Inc. 1425 The Deutsch Company Electronic Components Div. 1426 Haydon Switch Inc. 1427 Telerad Mfg. Corp. 1427A-1427B Magnetic Amplifiers, Inc. 1428-1430 Giannini Controls Corp. 1429 Valpey Crystal Corporation 1431-1433 Thompson Ramo Wooldridge Inc. 1432 Magnetic Metals Co.

(Continued on page 456.1)

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EE with 6 years experience in circuitry, transients, and transistor application. Will supervise development project on communications and frequency standard\$10-\$12,000

PROJECT ENGINEER

EE with 4-6 years experience with transistor circuity and knowledge of reaction of transistors at extreme environmental conditions ... to \$12,000

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PHYSICISTS

PROJECT LEADER—OPTICS

Head design and development of optical systems and components for infra-red detection to \$19,000

SENIOR CIRCUIT DESIGN

5 or more years experience in design. Develop and test transmitters, computers, and receiversto \$17,000

COMMUNICATION ENGINEERS

COMPUTER ENGINEERS

Digital to Analogue to \$16,000 Logical Design to \$15,000 Scientific Programmer ... to \$15,000

MANAGER

Radar receiver and microwave. Manage 4 engineering supervisors and 50 engineers. Designing radar antennas, microwave, coupling and receivers to \$20,000

APPLICATION ENGINEER

ELECTRONIC ENGINEER

PROJECT ENGINEERS

Either electrical or mechanical engineers with 7 to 10 years experience with emphasis on applied mechanics. precision instruments, radar, synchros and servo motors, missile research and developments, digital and analogue computer, etc. \$8-\$12,000

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World Radio History



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Your gateway to a systems engineering career is your experience in any of the following areas:

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Please reply fully in confidence to Mr. E. A. Smith, Box 3-F



452A

Whom and What to See at the **Radio Engineering Show**

(Continued from page 450A)

Zero Manufacturing Co., Booth 4118 1121 Chestnut St. Burbank, Calif.

Joe E. Daniels, Richard Belle Isle, Harry Halinton, Dan Greene, Lou Ackerman, Jerry Bouton, L. G. White, Robert Deunk

Complete line of packaging for the electronic industry including *new Zero Modular Re-Usable Shipping/Storage Container System, Deep Drawn Aluminum, Magnesium and Alply Transit Cases. electronic Zippertubing Co., Booth 4044 752 South San Pedro St. Los Angeles 14, Calif.

H. R. Edwards, Jack Mills, Lynd Walking, W. T. Murray, R. A. Murray

"Zippertubing" provides speedily applied pro-tection for cable and cable assemblies against heat, cold, abrasion, moisture, RF interference, chemical agents, etc. A wide variety of materials and types available.

NUMERICAL LISTING OF EXHIBITORS

Use this listing of exhibitors arranged in numerical order to help in interpreting the information given in the product listings which begin on page 468A.

A complete alphabetical listing of exhibitors, including the names of personnel manning the exhibits, and the products to be exhibited, begins on page 140A.

FIRST FLOOR

Main Lobby-The Institute of Radio En-

gineers Proceedings of the IRE The IRE Directory

Membership Information Service

- 1100 The Constanta Co. of Canada Ltd. 1100A Consolidated Resistance Co. of America, Inc.
- 1100B The Sibley Company
- 1101 Technical Appliance Corp.
- 1102 Fenwal Electronics, Inc.
- 1103 Avnet Electronics Corp.
- 1104 Autotronics Inc.
- 1105 Silicon Transistor Corp.
- 1106 Aircom, Incorporated
- 1107 Line Electric Co., Inc.
- 1108 Axel Brothers Inc.
- Electronics Div.
- 1109 Glass-Tite Industries, Inc.
- 1110 Applied Research Inc.
- 1111 Emerson & Cuming Inc.
- 1112 Columbia Technical Corp.
- 1113 ACF Industries, Inc. ACF Electronics Div.

- 1114-1115 Motorola, Inc. Semiconductor Products Div.
- 1116 Nicad Division
- Gould-National Batteries, Inc. 1117 Collins Electronics Mfg. Corp.
- 1118 A. W. Welch Manufacturing Co.
- 1119 Gordos Corporation
- 1120 Columbus Electric Mfg. Co.
- 1121 Three Point One Four Corp.
- 1201-1203 Helipot Div. Beckman Instruments, Inc.
- 1202 Dynamics Corporation of America Radio Engineering Laboratories, Inc.
- 1204-1206 The W. L. Maxson Corp.
- 1205 Shockley Transistor Corp. Sub. Beckman Instruments, Inc.
- 1207-1209 Diamond Antenna & Microwave Corp.
- 1208-1210 CBS Electronics
- 1211-1215 Burroughs Corporation
 - Electronic Tube Div.
- 1212-1214 General Transistor Corp.
- Standard Transformer 1216 Chicago Corp.

IN SAN DIEGO . . .

Electronics Engineering Opportunities

Unique civilian opportunities are available now for Electronic and Electrical Engineers at the U.S. Naval Repair Facility, San Diego. Field includes laboratory work, electro-magnetic propagation studies, and preparation of audio and visual amplifier and switching circuitry for complex communication and distribution systems.

Career Civil Service positions with salaries ranging from \$7510 to \$8230, and outstanding leave, insurance and promotion benefits.

Three years of professional engineering experience required, including one year of electronic engineering.

Submit resumes and inquiries to: Industrial Relations Department U.S. Naval Repair Facility San Diego 36. California . S. NAVY

⁽Continued on taac 454.4)

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doppler techniques has accounted for its rapid expansion to more than 375,000 square feet of the most modern facilities for research, development and production in San Diego and Torrance in Southern California. Here is your opportunity to combine challenging work, a bright future and life in an area famous for ideal year-round climate, excellent facilities for graduate work, fine schools for your children. For full particulars, send your resume or write for a brochure to Ryan Electronics, Dept. 1, 5650 Kearney Mesa Road, San Diego 11, California.

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ENGINEERS-SCIENTISTS

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Altitude—the missing third dimension, a means of target identity retention so vital in automatic tracking—will be provided by a new system, the Air Height Surveillance Radar, now under development by The W. L. Maxson Corporation. It also will aid traffic controllers in the identification of aircraft and the supervision of their altitudes as well supplement search scopes with altitude separation data and altitude layer filtering, all of which will improve both air safety and traffic handling capabilities.

Current engineering positions in other areas include: Radar Systems • High Power Radar Modulators • Product Design Microwave Antenna Design • Electronic Test Engineering Analog Computer Design • Microwave Research & Design Guidance Systems • Transistor Switching Circuits

Interviews Monday thru Friday, or send resume to: Mr. L. W. ALBRIGHT, Technical Placement Manager.



Whom and What to See at the Radio Engineering Show

(Continued from page 448A)

Yokogawa Electric Works, Inc. 40 Worth St. New York 13, N.Y. Booth 3054 Ken Sadohara, Toshi Sasaki, Karl A. Kopetsky, Yukio Horie, Bernard Fox



New MODEL TPF Thermo Milliammeter

Portable Precision Instruments: Model TPF Thermocouple Type HF Milliammeters & Voltmeters, covering up to 5 megacycles, Model MPF Moving Coil Type DC, Model SPF Moving Iron Type & Model CPF Rectifier Type AC Animeters & Voltmeters: Portable Insulation Tester: Model L-5: Vacuum Thermocouples.

York Metal Products, Inc., Booth 4239 350 Greenwich Street New York 13, N.Y.

Alvan Bisnoff, Steve Bucek, David Fox, Claude Bisnoff

Custom precision sheet metal fabrication, chassis, panels, cabinets, instrument cases, racks, subassemblies, angle iron and aluminum weldments. Bracket fabrication and assembly. Complete finishing facilities, baking enamels, hammertones and wrinkles.

Zagar, Incorporated, Booth 4241 24000 Lakeland Blvd. Cleveland 23, Ohio

John P. Mrsnik, Frank G. Zagar, Frank J. Zurga, Beverly D. Smith, Chas. M. Powers, Harry J. Hutter

Multiple spindle Quick Change Printed Circuit Drilling Equipment Drilling from 2 to 2,000 holes in one pass. Zagar Air Oil unit to illustrate and demonstrate method for drilling printed circuit boards.

Zell Products Corp., Booth 4108 276-80 Main St.

Norwalk, Conn.

Burton B. Zell, Sol Young, Russell Cameron, Don Bitko, Bradley Anschutz, Joe Smith, Al Seprinski, Bill Sansouci, Dean Lucas, Hal Castle, Jim Barnett, Art Hansen

Class to Metal Hermetic Seals, Kovar Stampings, Transistor Stems for Military and Industrial Applications (TO-5-TO-18), Multi Headers, Diodes, End Seals, Terminals, Connectors, Matched and Compression type plated to required finish-to pass CP4, Mandrel Bake and Solderability Tests.

Zenith Radio Corporation 6001 Dickens Ave. Chicago 39, Ill. Booth 3948

▲ Ed Dervishian, Scott Morency, Tom Storey, Don Anderson, ▲ William Van Slyck

Electron beam parametric amplifier and equipment to demonstrate 1 db noise figure. Portable secondary time standard. Central Electronics 100 v single sideband transmitter, Flash X-Ray Equipment.

(Continued on page 452A)



Westinghouse-Baltimore has a radar story to tell you. For a meaningful engineering career, you owe it to yourself to know what the Baltimore Divisions of Westinghouse are doing...and how you might participate in exciting programs now under way. We welcome the opportunity to give you the facts...and let you judge for yourself. If you are an experienced engineer, learn about our advanced development of new weapons systems involving revolutionary data processing, antenna systems, and system synchronization. These offer immediate career opportunities for both electrical and mechanical engineers with experience or interest in the following fields:

Molecular Electronics Microwave Systems Advanced Antenna Systems Liaison and Field Engineering Transmitter and Receiver System Development System Synchronization Digital Computer Design Advanced Data Displays Systems Reliability Prediction Advanced Systems Automated Test Mechanical Design & Packaging New York Interviews While attending the I. R. E. Show, be sure to visit us at the Con. vention Hotel, or call MUrray Hill 8-0163.

Send resume to: Mr. A. M. Johnston, Dept. 262


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Several supervisory positions open for men of exceptional background and ability

Also Positions open for: PHYSICISTS – ELECTRONIC ENGINEERS

If you qualify for a career at Hoffman, you will be a part of the world's most experienced team in silicon semiconductor technology. Among this division's achievements are the world's first commercial silicon diode, silicon zener diode and silicon solar cell. In addition, the division makes the most extensive line of silicon semiconductor devices in the industry – for both military and commercial markets. At Hoffman you will command every facility you need to carry out your new ideas. And you will be part of a *growing* company – which means that you can build your future as fast as your talents permit. U. S. citizenship not required.

Please submit resume to W. L. Parker, Personnel Manager.

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Unique mesa configuration of Hoffman 2N696, 2N697 silicon transistors, conceived by Hoffman Semiconductor R & D staff, increases usefulness of emitter area, boosts high-frequency power gain to three times that of similar devices.



Whom and What to See at the Radio Engineering Show

(Continued from page 445A)

Winchester Electronics, Inc., Booths 2121-2123

19 Willard Road Norwalk, Conn.

A Donald R. De Tar, L. E. Harrod, E. C. Quackenbush, John Lynch, Herbert Sherman, Larry Bordeau, Frank Cowe, Louis Bencze, James Murphy, Joseph Roos, Al Octavio, Robett Thornley

Electronic Connectors: Crimp Type-Removable Contact, Hermetics, Resilient Inserts, A.N. Types, Environmental Types, Crimping Tools, Relay Sockets, Tube Sockets, Terminals, Printed Circuit Connectors, Rack and Panel, Sub Miniatures, MIL Standard Types, Test Receptacles, Hoods, Shells, Potting Forms, Special Custom Designs.

Wind Turbine Co., Booths 1712-1714 248 E. Market St. West Chester, Pa.

Robert W. Weeks, Kenneth B. Havens, ▲ Albert C. Veldhuis, Davis B. Oat, Peter G. Park, Charles K. Hutchison, Fred H. Lukens Trylon Towers and Antennas.

Wright Metalcoaters, Inc. 255 West St. South Hackensack, N.J. Booth 4007

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(Continued on page 450A)

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Whom and What to See at the Radio Engineering Show

(Continued from page 442A)

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(Continued on page 448A)

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 440.4)

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Roger White Electron Devices, Inc., Booths 1610-1618, 1709-1717 See: Litton Industries, Inc.

Whittaker Gyro, Booth 2128 See: Telecomputing Corp.

(Continued on page 445A)

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PROCEEDINGS OF THE IRE March, 1960

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Whom and What to See at the Radio Engineering Show

(Continued from page 439A)

Weller Electric Corporation, Booth 4024 601 Stones Crossing Rd. Easton, Pa

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Western Gold & Platinum Co., Booths 4201-4203 See: Wilbur B. Driver Co.

Western Lithograph Co., Booth 4402 See: Westline Products Div.

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See: Union Switch & Signal Div.



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(Continued on page 442.1)

Whom and What to See at the Radio **Engineering Show**

(Continued from page 436.4)

Wayne Kerr Corporation, Booths 3827-3829

1633 Race St.

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▲ Boyce M. Adams, Gordon L. Ball, John Robertshaw, Jr., John Gruenberg II

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(Continued on page 440A)

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PROCEEDINGS OF THE IRE March. 1960



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Whom and What to See at the Radio Engineering Show

(Continued from page 434A)

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(Continued on page 439A)

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 432A)

Wallson Associates, Inc. 912 Westfield Ave. Elizabeth, N.J. Booth 3006

▲ Joel P. Wallenstein, Abram S. Jaffe, H. J. Goldstein, ▲ Gerald Randolph, Marvin Elkin-son, Gerard Dravis, Patrick Marshall



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(Continued on page 436A)



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(Continued from page 428A)

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See: Nems-Glarke Co.

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Norman F. Weyland, Joseph L. Stella, William A. Schrader, Arthur H. Strickland, Robert H. Tremble, David Flint, Joseph J. Miranto, Edward W. Cassidy, Bruce W. Cameron, Adrian W. Doherty, Russell A. Johnson, Arthur K. Schott

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(Continued from page 427A)

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(Continued on twie 43221)

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(Continued from page 424A)

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	ADDRESS				
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Whom and What to See at the Radio **Engineering Show**

(Continued from page 422A)

Universal Mfg. Co., Inc. 1168 Grove St. Irvington 11, N.J. Booths 4415-4417 William A. Bernau, Anthony O. Vicari, Richard S. Hillebrand, A Melvin E. Liberman



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Universal Transistor Products Corp., Booths 3612-3614 See: Telechrome Mfg. Corp.

Universal Winding Co., Booths 4323-4325

See: Leesona Corporation

Utica Drop Forge & Tool Div., Kelsey-Hayes Co., Booth 4002

F. L. Marshall, F. J. Stiefvater, W. I. Pugh, H. Neff, C. Ellingwood, R. C. Bryan, L. T. Bryan, R. Dunn, A. Kuflan, W. Rozmus, E. Munson

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Utrad Corp., Booths 1610-1618, 1709-1717

See: Litton Industries, Inc.

Valpey Crystal Corporation 1244 Highland St. Holliston, Mass. Booth 1429

T. S. Valpey, Jr., Nelson B. Piper, Nor-man R. Gillin, R. S. Puleo

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D. Van Nostrand Co., Inc., Booth 4429 120 Alexander St.

Princeton, N.J.

David Phillips, Peter H. Fagnano, Mulford Colebrook, Dayton Beguelin, E. M. Crane, Jr., Adrian Clark

The D. Van Nostrand Publishing Company will display reference books and publications on radio engineering, electronics and communications, and related subjects.

(Continued on page 427A)

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March, 1960

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Semiconductor Products Department, Electronics Park, Syracuse, N.Y.

Whom and What to See at the Radio **Engineering Show**

(Continued from bage 420A)

United Transformer Corp. 150 Variek St. New York 13, N.Y. Booths 2413-2414 Hank Russell, Ted Craige, Joe Barreca, Austin Profeta, Mike Cooney, Walt Rooney, Bob Soevyn



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Unitek Corp., Booth 4526 See: Weldmatic Division

Universal Controls, Inc., Booths 2218-2220

See: C. P. Clare & Co.

Universal Electronics Co., Booth M-11 1720 Twenty-second St. Santa Monica, Calif.

Edward Lacey, Eugene Cahn, William Fried-man, Robert Schermerhorn

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Universal Instruments Corp., Booths 4019 & Room 610 139 East Frederick St.

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(Continued on page 424A)

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providing complete information on the firms providing any specific product or service is available from the information booth at the head of the escalators on the third floor of the Coliseum.



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IRE Show interviews — Collins will be interviewing in New York, Monday, March 21 through Thursday, March 24. For a personal, confidential interview, phone Mr. L. R. Nuss, PLaza 5-4580. A convenient appointment time will be arranged. If unable to interview at this time, send your resume to: Mr. L. R. Nuss, Collins Radio Company, Cedar Rapids, Iowa; Mr. Ben E. Jeffries, Collins Radio Company, 1930 Hi-Line Drive, Dallas 7, Texas; or Mr. F. W. Salyer, Collins Radio Company, 2700 W. Olive Ave., Burbank, California,



COLLINS RADIO COMPANY . CEDAR RAPIDS, IOWA . DALLAS, TEXAS . BURBANK, CALIFORNIA

421A



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Whom and What to See at the Radio Engineering Show

(Continued from page 418A)

U.S. Semiconductor Products, Inc. Div. of United Industrial Corp. 3540 West Osborn Rd. Phoenix, Ariz. Booths 2713-2715

William R. White, ▲ J. C. Worth, Jr., ▲ Robert R. Rutherford, ▲ Edward Botwinick, R. Hales Pridmore, Milton I. Liebhaber



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U.S. Stoneware Co. Alite Division Orrville, Ohio Booths 2205-2207

J. M. W. Chamberlain, Howard Farkas, II. Frahme, W. E. Coykendall, Jr., Walter Perkins, Walter Tarnacki, John F. Schuck



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United States Time Corporation, Booth 3841

375 Park Ave. New York 22, N.Y.

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(Continued on page 422A)

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Electronic Systems Division





VANIA Dept. B3, Box 188, Mountain View, California

PROCEEDINGS OF THE IRE March, 1960

World Radio History



U. S. Navy P5M-2 antisubmarine patrol seaplane, produced by Martinequipped with TI-built AN/APS-80 surface search radar, AN/APA-125A radar indicator, AN/ASQ-8 magnetic anomaly detector and TD-239A intervalometer.

TI RADAR & MAGNETICS IN ANTISUBMARINE SYSTEMS



TEXAS INSTRUMENTS



Whom and What to See at the Radio Engineering Show

(Continued from page 416A)

United Mineral & Chemical Corp. 16 Hudson St. New York 13, N.Y. Booth 1627B A Irwin Stelzer, Herbert Rosenthal, Alexander Imich, Terry Koncelik, Ed Wiest, Anita Rosenthan, Manfred DeRewal



Resistors

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U.S. Ceramic Tile Co., Booth 4109 See: Diamonite Products Mfg. Co.

> U.S. Components, Inc. 454-462 East 148th St. New York 55, N.Y. Booth 2805

▲ B. A. Jackson. Henry Nalbantian, Steve Nalbantian, Erncst Klinger, B. R. Remondino



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U. S. Engineering Co., Booths 1610-1618, 1709-1717 See: Litton Industries, Inc.

See: Lation Industries, Inc.

U. S. Gasket Co., Booths 2814-2816 See: Garlock Electronic Products

U. S. Semcor, Booths 2713-2715 See: U.S. Semiconductor Products, Inc.

(Continued on page 420A)

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World Radio History



Hughes' FALCON Air-to-Air Missiles in front of Convair F-102A

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 414A)

Union Carbide Consumer Products Co., Div. Union Carbide Corporation, Booths 2401-2403

30 East 42nd St.

New York 17, N.Y.

New Fork 17, N.F. C. Anderson, D. B. Ashway, C. P. Barry, W. A. Bruce, R. S. Burgess, ▲ D. B. Came-ron, A. F. Carey, H. E. Carpenter, S. I. Con-verse, ▲ H. R. Erskine, C. R. Fisher, W. S. Gillette, H. J Harlow, R. F. Kiefer, F. A. Langell, ▲ D. R. Ogden, F. B. Pipal, ▲ N. M. Potter, C. J. Sullivan, D. G. Taylor, D. P. Trepte, R. A. Varsha, S. M. Wall

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Union Carbide Corp., Booth 2405 See: Kemet Company

Union Switch & Signal Div., Westinghouse Air Brake Co., Booths 2122-2124 Pittsburgh 18, Pa.

Paul K. Eckhardt, H. J. Myers, K. E. Doriot, A. E. Over, G. A. Dawes, J. W. Hansen, F. E. Baxter

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Chicago 7, Ill.

Royce Young, Larry Gernert, Ed Salisbury

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United-Carr Fastener Corp., Booths 2535-2536

31 Ames St.

Cambridge, Mass.

See: Cinch Mfg. Company and Ucinite Co.

United Catalog Publishers, Inc., Booth 4105

60 Madison Ave.

Hempstead, N.Y.

Arthur I. Rabb, Samuel Roth, A. E. Stevens, George Siegel, Harry Birse, George Kerner, A Irving J. Frisch, Robert J. Males, Ray Smyth, Curtis E. Glanville, John Mitchell, Harold Gabriel

Harold Gabriel 1960 edition—EEM—Electronic Engineers Mas-ter, Catalog-Directory of the industry, 1960 edi-tion The Radio-Electronic Master, Catalog of standard products sold by electronic parts dis-tributors, Electronic Products Magazine—the industry's only new product monthly. File-O-Matic, perpetually up-to-date catalog of standard electronic products. Pricing Service—prices of standard electronic products. Audio-File—catalog-pricing service for audio hi-fi products.

United Industrial Corp., Booths 2713-2715

See: U.S. Semiconductor Products, Inc.

(Continued on page 418A)

Show Hours 10 a.m. to 9 p.m. daily Monday through Thursday March 21-24, 1960



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World Radio History



19

Anti-personnel Mine Detector AN/PRS-3 (XR-12) designed and built for the Corps of Engineers by Texas Instruments.

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 410A)

George Ulanet Co. 413 Market St. Newark 5, N.J. Booth 2119

H. Ulanet, A. W. Burke, George Ulanet, James K. Dennis



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Unimax Switch Division, The W. L. Maxson Corp., Booths 1204-1206 Ives Road

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(Continued on page 416A)

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ENGINEERS

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Whom and What to See at the Radio Engineering Show

(Continued from page 404A)

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USECO, Inc., Booths 1610-1618, 1709-1717

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(Continued on page 414A)

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World Radio History

working

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machine's decision-space

man's decision-place



living

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2 Radar Equipment Systems Specialist: This position calls for a creative engineer capable of conceiving and directing the design of long-range radar systems. Desirable experience includes around ten years in

B Advanced Systems Engineer: This position calls for a creative engineer capable of defining future defense and space detection problems as well as the ability to conceive and establish the feasibility of optimum systems solutions to these problems—making use of the most advanced techniques and understanding. He must recognize the need for and coordinate the development of new techniques and the exploration of

Advanced Radar Systems Analysis and Development Engineer: Engineers are needed who are able to visualize and define future defense and space problems—conceive advanced radar systems to solve them. An advanced degree and/or strong background in system analysis and design is essential. Assignments open

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at least one of the following: radar systems design, antenna systems, R.F. components, radar receiver systems or radar data processing systems. Salary structure is equal to the challenge.

new phenomena in the area of detection systems. Background desired: Bachelor degree plus a combination of advanced training and several years experience in both the theoretical and practical aspects of detection systems engineering. A desire to work in the conceptual phase of system design with the analytical ability required to evaluate and demonstrate the effectiveness of proposed systems.

include: analyze and define requirements for advance detection systems and determine broader parameters for such systems, establish their feasibility; analyze long range missile detection systems and specify optimum configuration on the basis of utility, performance, cost and delivery. 228-9

Write in confidence to T. M. George, Supervisor—Personnel Administration

Missile Detection Systems Section HEAVY MILITARY ELECTRONICS DEPARTMENT



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World Radio History

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Whom and What to See at the Radio Engineering Show

(Continued from page 402A)

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Whom and What to See at the Radio Engineering Show

(Continued from page 400A)

Transitron Electronic Corp. 168 Albion St. Wakefield, Mass. Booths 1319-1323 H. Thomas Neavitt, Lawrence W. King, ▲ William Slusner, ▲ Nick De Wolf, Charles Hill, Peter Jenner, Thomas Clark



Silicon Controlled Rectifier (Left) The Transwitch, Transistor Package (Right)

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Triad Transformer Corp., Booths 1610-1618, 1709-1717

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Trio Laboratories, Inc. DuPont Dr. Plainview, L.I., N.Y. Booth 3033 S. Salz, Jurgen Worthing, Philip (

▲ Jay S. Salz, Jurgen Worthing, Philip Greenstein, John R. Crawford, Harold D. Miller



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(Continued on page 404.4)

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402A

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Whom and What to See at the **Radio Engineering Show**

(Continued from page 398A)

Times Facsimile Corp., Booths 1610-1618, 1709-1717 See: Litton Industries, Inc.

Tinnerman Products, Inc., Booths 4207-4211

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(Continuea on page 402A)



400 A

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Whom and What to See at the Radio Engineering Show

(Continued from page 397A)

Thomas & Skinner, Inc., Booth 2929 1120 East 23rd St.

Indianapolis 7, Ind.

James C. Skinner, Stephen J. Gavin, Robert Fulton, Richard Hansen, Edward Cronk, Rollin J. Millar

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Thompson Products, Booths 1431-1631 See: Thompson Ramo Wooldridge Inc.

Thompson Ramo Wooldridge Inc.

8433 Fallbrook Ave., Canoga Park (Los Angeles), Calif. and 23555 Euclid Ave., Cleveland 17, Ohio Booths 1431-1631

Booths 1431-1631 Thompson Ramo Wooldridge Inc.: Charles Wacker; DAGE Television Division: O. D. Page, S. T. Spangler, D. A. Schonmeyer; Ramo-Wooldridge Division: Gary Langseth, Milton Baldridge, Ross Penny, Ken Busche, Harold Luxenberg, Denny Pidhayny, Lawrence Murdock, Roger Trapp, Richard Nishimura; TAPCO Group: F. G. Weihmiller, J. N. McCarthy, R. T. Meyers, W. W. Welsh, P. A. Trostel, W. H. Baucom, C. W. Chase, T. B. Ray, C. W. Sargent; Thompson-Ramo-Wool dridge Products Co.: Ken Hayes, James Carolan, Bertram Newman, Henry Bechard, Lewis Ward, F. Lee Johnson, Dan L. McGurk



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Three Point One Four Corp. 24 Woodworth Ave. Yonkers, N.Y. Booth 1121 J. J. Root, Richard Sapolin, Michael Odlivak, George Sokos



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(Continued on page 400.4)

Whom and What to See at the Radio Engineering Show

(Continued from page 394A)

Texas Instruments Incorporated Semiconductor-Components Division 13500 North Central Expressway Dallas, Tex.

Booths 1409-1421

Mark Shepherd, Jim Carland, Harry Owens, Jim McDade, Jay Reese, Ed Brierty, Jess Moore, Bob Votteler, Leonard Maguire, Jim Brown, Steve Karnavas, George Deaderick, Dick Hanschen, Jack Ohlrich, Charley Clough



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Textron, Inc., Booths 2126, 3920-3922, 3107-3109

See: G-C Electronics Mfg. Co., California Technical Industries, MB Electronics Div.

Thermo Electron Engineering Corp. 127 Smith Place Cambridge 38, Mass. Booth 3950 A Lawrence T. Sullivan, Gabor Miskolczy, Lazarus Lazaridis, John A. Weish, L. Clifford Schroeder, Thomas A. Robinson, Peter G. Witherell First Public Demonstration—*Thermionic Converter, converts heat directly into electricity—Not for Sale.

Thomas & Betts Co., Inc., Booth 1729 36 Butler Street

Elizabeth, New Jersey E. D. Thomas, D. J. Crimmins, M. C. Logan, R. Grant, D. Lipps, F. Dixon, G. McGrane, J. O'Donnell, V. J. Brennan, P. Garippa, B. Nehring, J. Craig

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Thomas Electronics, Inc., Booth 2913 118 Ninth St.

Passaic, N.J.

Robert G. Scott, I. J. Posner, Jess E. Dines, John Loschiavo, Gerald Cornell, \blacktriangle N. Broderick, \blacktriangle K. A. Hoagland, \bigstar E. Lisovicz, P. Nuccio, \blacktriangle P. Seats, L. Connelly, M. Beasty 23" and 19" television picture tubes; low voltage aluminized instrument tubes; flying spot scanner tubes; monitor tubes; industrial-military and special-purpose cathode ray tubes; electronic tube components; cable assemblies and wire harnesses; test equipment.

(Continued on page 398A)

▲ Indicates TRE member. * Indicates new product.





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World Radio History

Ivan R. Samuels Director of Personnel

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396A

March, 1960



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By Armed Forces Veterans

(Continued from page 392.4)

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 380A)

Texas Instruments Incorporated Geosciences & Instrumentation Division 3609 Buffalo Speedway Houston 6, Tex.

Booths 1409-1421

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(Continued on page 397.4)

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World Radio History

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By Armed Forces Veterans

(Continued from page 390A)

radar systems. Some teaching and research experience. Former Signal Corps officer. Liceused Professional Engineer in New York. Box 2059 W

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Desires position not wholly technical involving possible overseas travel. LTJG USNR. Tau Beta Pi, Eta Kappa Nu. Unmarried, BS, and MS, in E.E. from large midwestern universities. Box 2060 W.

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ntinued on page 391.1)



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and on page soury

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960



(Continued from page 384A)

uate programs in engineering, developing undergraduate and later graduate programs in sciences, constructing new science and engineering building to prepare engineers for the industrial and technical development of Turkey and the Middle East. Address inquiries to Dean Howard P. Hall, School of Engineering, or Prof. Frank Potts. Acting Dean, School of Sciences, Rohert College, Bebek, Post Box 8, Istanbul, Turkey, with copy to Near East College Assoc., 40 Worth St., Room 521, New York 13, N. Y.

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(Continued from page 382A)

MICROWAVE SPECIALIST

Microwave physicist needed for applying microwave techniques to the study of plasma flows and ionized regions around high speed models and in shock tubes. Measurements and study of the radio-frequency energy emitted by the passage of high-speed models and of the transmission and reflection characteristics of the wake are required in order to evaluate the effects of these characteristics and also as an aid to further the knowledge of flow phenomena at extreme speeds, Applicant should have advanced degree with a good background in microwave propagation and field theory as well as ability to work with microwave hardware. He should be capable of taking the initiative in the application of microwave techniques and in the interpretation of results. Write Personnel Officer, NASA, Ames Research Center, Moffett Field, Calif.

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(Continued on page 387A)



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(Continued on page 384A)

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Whom and What to See at the Radio Engineering Show

(Continued from page 378A)

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L. Barton, C. Beran, E. Caroe, D. Dombrow-sky, W. Euerle, G. Heller, F. Marocco, G. Melody, E. O'Neill, R. Ruhling, W. Samson-off, D. Tornello, W. Wrobleski



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Oscillogram showing drive pulse and switching time for TM 301-02 core. Time scale $0.02\mu s/division.$

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(Continued from page 376A)

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(Continued on page 380A)

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F 3 Filament 90018 TF4RX01FB004 F 4 Filament 90019 TF4RX01FB006 F 5 Filament 90020 TF4RX01FB006 F 6 Filament 90021 TF4RX01FB006 F 7 Filament 90021 TF4RX01FB006 F 8 Filament 90021 TF4RX01B002 F 8 Filament 90023 TF4RX01B0012 F 9 Filament 90024 TF4RX01B012 F 10 Filament 90025 TF4RX01KB0012 F 10 Filament 90025 TF4RX01KB012 Direct Operation from Line Voltage Fast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 103 Motor	GF 2	Filament	90017	TF4RX01GB003
F 4 Filament 90019 TF4RX01HB003 F 5 Filament 90020 TF4RX01FB006 F 6 Filament 90021 TF4RX01FB006 F 7 Filament 90021 TF4RX01B003 F 8 Filament 90022 TF4RX01B002 F 9 Filament 90023 TF4RX01B002 F 9 Filament 90024 TF4RX01B002 F 10 Filament 90025 TF4RX01B012 F 10 Filament 90025 TF4RX01KB013 MAGNETIC AMPLIFIERS AMIL Specifications No Tubes Direct Operation from Line Voltage F ast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10 ^a Power Gain 2 x 10 ^a Fransistor Mag. Amp. Motor MAF-5 MAT-1 Wt. 18 oz. Wt. 18 oz. Wt. 10 oz. Ft6. 2 Ft6. 4	GF 3	Filament	90018	TF4RX01FB004
F 5 Filament 90020 TF4RX01FB000 F 6 Filament 90021 TF4RX01G8003 F 7 Filament 90021 TF4RX01B008 F 8 Filament 90021 TF4RX01B002 F 9 Filament 90021 TF4RX01B002 F 9 Filament 90024 TF4RX01B002 F 10 Filament 90025 TF4RX01B002 No Tubes Direct Operation from Line Voltage Direct Operation from Line Voltage Poser Gain 2 x 10 ² Power Gain 2 x 10 ² F Preamp. Maf-5 Motor Preamp. MAF-5 Motor MAT-1 Wt. 18 oz. Wt. 18 oz. Wt. 10 oz. F E E	GF 4	Filament	90019	TF4RX01HB005
Fr definition 90021 TF4RX01G8000 FF 7 Filament 90022 TF4RX01J8008 FB Filament 90021 TF4RX01J8008 F9 Filament 90021 TF4RX01J8001 F9 Filament 90021 TF4RX01J8001 F10 Filament 90021 TF4RX01J8001 F10 Filament 90025 TF4RX01J8001 MAGNETIC AMPLIFIERS • Hermetically Sealed To MIL Specifications No Tubes • Direct Operation from Line Voltage • Fast Response • Long Life Trouble Free Operation • Phase Reversible Output • • Power Gain 2 x 10 ³ • <p< td=""><td>AGF 5</td><td>Filament</td><td>90020</td><td>TF4RX01FB006</td></p<>	AGF 5	Filament	90020	TF4RX01FB006
F.F. 7 Filament 90022 TF4RX01JB002 F.F. 8 Filament 90023 TF4RX01KB001 F.9 Filament 90023 TF4RX01KB003 F.9 Filament 90023 TF4RX01KB003 F.9 Filament 90025 TF4RX01KB013 F.10 Filament 90025 TF4RX01KB013 MAGENETIC AMPLIFIERS • Hermetically Sealed To Mil Specifications • No Tubes • Direct Operation from Line Voltage • Fast Response • Long Life Trouble Free Operation • Power Gain 2 x 10 ³ • Power Gain 2 x 10 ³ • MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. Ff0. 2	AGF 6	Filament	90021	TF4RX01GB007
F B Filament 90023 TF4RX01R800X F 9 Filament 90024 TF4RX01B012 F 10 Filament 90025 TF4RX01R801X MAGNETIC AMPLIFIERS • Hermetically Sealed To Mill Specifications • No Tubes • Direct Operation from Line Voltage • Fast Response • Long Life Trouble Free Operation • Phase Reversible Output • Power Gain 2 x 10 ⁸ • Matrix • Pice 2 • Pice 4	GF 7	Filament	90022	TF4RX01JB008
A V Fildment 90024 IP4KA0136012 F 10 Fildment 90025 TF4RX01KB013 MAGNETIC AMPLIFIERS Hermetically Sealed To Mil Specifications No Tubes Direct Operation from Line Voltage Fast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10 ³ Fransistor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. Fr0. 2	GF B	Filament	90023	TF4RX01KB009
MAGNETIC AMPLIFIERS • Hermetically Sealed To Mil Specifications • No Tubes • Direct Operation from Line Voltage • Fast Response • Long Life Trouble Free Operation • Phase Reversible Output Power Gain 2 x 10 ³ • Mars.	GF V	Filament	90024	TEADYOL VB012
Specifications No Tubes Direct Operation from Line Voltage Fast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10 ^a Power Gain 2 x 10 ^a Transistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. FIG. 2	M	AGNETIC	AMPL	IFIERS
 No Tubes Direct Operation from Line Voltage Fast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10⁸ Power Gain 2 x 10⁸ 	M	AGNETIC		IFIERS
 Direct Operation from Line Voltage Fast Response Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10⁸ Power Gain 2 x 10⁸ Transistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. FiG. 2 	M H s	AGNETIC ermetically pecifications	AMPL Sealed To	IFIERS MIL
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 Long Life Trouble Free Operation Phase Reversible Output Power Gain 2 x 10^a Power Gain 2 x 10^a Fransistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. 	M • H • S	AGNETIC lermetically pecifications lo Tubes irrect Operat	AMPL Sealed To ion from	IFIERS MIL Line Voltage
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Fransistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz.	M H S D F F	AGNETIC lermetically pecifications to Tubes tirect Operat ast Response ong Life Tra	AMPL Sealed To ion from	IFIERS MIL Line Voltage Operation
Transistor Preamp. MAT-1 Wt. 18 oz. Wt. 10 oz.	M • H S • N • D • F • L • P	AGNETIC Intermetically pecifications to Tubes irrect Operat ast Response ong Life Tro hase Reversi	AMPL Sealed To ion from uble Free ble Output	Line Voltage
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Transistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz. Fric. 2	M • H • S • N • D • F • L • P	AGNETIC pecifications lo Tubes birect Operat ast Response ong Life Tro hase Reversi Power	AMPL Sealed To ion from uble Free ble Outpu Gain 2 x	Line Voltage Operation
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Transistor Mag. Amp. Motor Preamp. MAF-5 MAT-1 Wt. 18 oz. Wt. 10 oz.	M • H • C • C • C • C • C • C • C	AGNETIC lermetically pecifications to Tubes birect Operat ast Response ong Life Tro hase Reversi Power	AMPL Sealed To ion from uble Free ble Outpu Gain 2 x	LIFIERS MIL Line Voltage Operation V1 10 ²
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Laad Laad	M • H S • D • F • L • P • P • Trans • Prea MA [*] Wt. From	AGNETIC Permetically 3 pecifications to Tubes to Tubes ong Life Tro hase Reversi Power Fower istor Ma mp. 1 T-1 Wr 10 oz.	AMPL Sealed To ion from uble Free ble Outpu Gain 2 x Gain 2 x g. Amp. MAF-5 . 18 oz.	IFIERS MIL Line Voltage Operation 102 Motor
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40 0 20 10	MA FIC	AGNETIC Jermetically 3 pecifications to Tubes to Tubes ong Life Tro hase Reversi Power Jone Composition power Agent Agent Agent power Agent Agent Agent power Agent Agent Agent Agent Agent Agent Agent Agent Agent Agent Ag	AMPL Sealed To ion from uble Free ble Outpu Gain 2 x	Line Voltage Operation U1 10 ³ Motor

TORC	DIDAL II	NDUCTORS		HE	RMETIC	ALLY S	EALE	D,
MIL Grade 4 — Metal Case MIL Grade 5 — Malded Hagged Units				PU	LSE TR	ANSFO		RS imum
Uncased Units Highert 0				pulse	performance	e.	una opi	mom
	Highest Q			For	use in bloc	king oscillat	tor, inte	rstage
•	Highest self	resonant freq.		coup	ling and lov	v level outp	ut circuit	ts.
1.60	Law temper	ature coefficient		Rugg	edized const	ruction — C	Grade 4.	
	No hum pick	up-astatic construction		• Serie	s or paralli	er connection	n or will	naings
	Can be supp	olied with center taps		DM-		8 DM	-20	F. DM-
		1000		Cot No.	MIL Type	Pulse Volt Kilovolt	oge Cl	or. Im Ohms
				MPT- 1	TE4RX35YY	0 25 0.25	0.25	250
FREQL	JENCY RANGE:	500CP TO 15KC		MPT- 2	TF4RX35YY	0.25 0.2	25	250
Туре	Max Q	Inductance Range		MPT- 3	TE4RX35YY	05050	0.5	250
11-11	290	TMH to 50Hy		MPT. 4	TEARX35YY	05.05	5	250
TI-12	255	1MH to 30Hy		METO 9	TEADYDEVY	0505	3.5	500
TI-1A	230	5MH to 20Hy		MPT- 5	TP4RAJJIT	03030		500
TI-4	195	5MH to 5Hy		MPT- 6	IF4RA35TT	0 5 0.5	>	500
TI-5	1 30	5MH to 2Hy		MPT- 7	TF4RX35YY	070.70	0.7	200
TI-16	72	1 MH to 2Hy		MPT- 8	TF4RX35YY	0707	7	200
FREQ	UENCY RANGE	· 10KC TO 50KC		MPT- 9	TF4RX35YY	10101	1.0	200
71.12	202			MPT-10	TF4RX35YY	1.0 1 0		200
TL-7	303	IMH to 500MH		MPT-11	TF4RX35YY	10101	1.0	500
TI-6	279	1MH to 400MH		MPT-12	TF4RX35YY	0.15 0.15 0.	.3 0.3	700
Ť1-7	200	.500MH to 200MH		Statement of the local division in which the local division in the	State Lands		-	and the second
T1-17	110	.100MH to 100MH			and the second second		-	-
FREQU	JENCY RANGE	30KC TO 200KC				NEMI		
					/			
TI-19	115	1 MH to 100 MH						
TI-18 TI-8	115	.1MH to 100MH		н	RMFTI	CALLY	SFAL	FN
TI-18 TI-8 TI-10	115 140 185	.1MH to 100MH .1MH to 100MH 1MH to 200MH		H	RMETI	CALLY	SEAL	ED
TI-18 TI-8 TI-10 TI-9	115 140 185 175	.1 <u>MH to 100MH</u> .1 <u>MH to 100MH</u> 1 <u>MH to 200MH</u> 1 <u>MH to 500MH</u>		HI	ERMETI CONSTA	CALLY NT VOI	SEAL LTAGI	ED E
TI-18 TI-8 TI-10 TI-9 TI-19	115 140 185 175 100	1MH to 100MH 1MH to 100MH 1MH to 200MH 1MH to 200MH 1MH to 500MH 1MH to 5MH		HI	CONSTA	CALLY NT VOI	SEAL LTAGI	ED E
TI-18 TI-8 TI-10 TI-9 TI-19 TI-3	115 140 185 175 100 260	.1MH to 100MH .1MH to 100MH 1MH to 200MH 1MH to 500MH .1MH to 50MH .1MH to 10MH		HI	ERMETIC CONSTA TRAN	CALLY NT VOI SFORME	SEAL Ltagi Ers.	ED E
TI-18 TI-8 TI-10 TI-9 TI-19 TI-3 TI-3A	115 140 185 175 100 260 310	.1MH to 100MH .1MH to 100MH 1MH to 200MH 1MH to 200MH .1MH to 50MH .1MH to 5MH .1MH to 10MH 10MH to 100MH	ľ	HI (CONSTA TRANS	CALLY NT VOI SFORME	SEAL LTAGI ERS.	ED E
TI-18 TI-8 TI-10 TI-9 TI-19 TI-3 TI-3 TI-3A	115 140 185 175 100 260 310 GH FRE	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH .1 MH to 500MH .1 MH to 50MH .1 MH to 10MH 10MH to 100MH QUENCY		HI Meets Specifi No Tu	CONSTA CONSTA TRANS Military cations bes	CALLY NT VOI SFORME Accure Fast F	SEAL LTAGI ERS. ate Reg Respons	ED E Julatio
TI-18 TI-8 TI-9 TI-10 TI-9 TI-19 TI-3 TI-3A HIC TORC	115 140 185 175 100 260 310 GH FRE	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 500MH .1 MH to 500MH .1 MH to 100MH 10MH to 100MH QUENCY NDUCTORS		Meets Specifi No Tu No Ma	CONSTA TRANS Military cations bes oving Parts	CALLY NT VOI SFORME Accure Fast F Fully	SEAL TAGI RS. ate Reg Respons Automa	ED E Julatio e atic
TI-18 TI-8 TI-10 TI-9 TI-19 TI-3 TI-3A HIC TORC	115 140 185 175 100 260 310 GH FRE	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH .1 MH to 200MH .1 MH to 50MH .1 MH to 50MH .1 MH to 100MH 0 UENCY NDUCTORS		Meets Specifi No Tu No Ma	CONSTA CONSTA TRANS Military cations bes oving Parts	CALLY NT VOI SFORME Accure Fast F Fully	SEAL TAGI RS. ate Reg Respons Automo	ED E ulatio e atic
TI-18 TI-8 TI-10 TI-10 TI-9 TI-19 TI-3A TI-3A HI0 TORC FREQI	115 140 185 175 100 260 310 GH FRE DIDAL I	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 200MH .1 MH to 500H .1 MH to 50MH .1 MH to 10MH 10 MH to 100MH QUENCY NDUCTORS : 20KC TO 10MC		Meets Specifi No Tu No Ma	Military cations bes oving Parts	CALLY NT VOI SFORME Fast F Fully	SEAL TAGI ERS. ate Reg Respons Automo	ED E Julatio e atic
T1-18 T1-8 T1-10 T1-19 T1-3 T1-3A HI0 FREQI T1-21	115 140 185 175 100 260 310 GH FRE DIDAL I UENCY RANGE 205	.1 MH to 100MH .1 MH to 100MH .1 MH to 200MH 1 MH to 500MH 1 MH to 500MH .1 MH to 500MH .1 MH to 500MH .1 MH to 100MH .1 MH to 100MH .1 MH to 100MH .1 MH to 100MH .1 0 MH to 100MH .1 0 MH to 100MH .1 0 MH to 100MH .2 0 KC TO 10 MC .01 0 MH to 150MH		Meets Specifi No Tu No Ma	Military cations bes oving Parts	CALLY NT VOI SFORME Fast F Fully	SEAL LTAGI ERS. ate Reg Respons Automa	ED E Julatio e atic
T1-18 T1-8 T1-10 T1-10 T1-9 T1-3 T1-3A HIC TORC FREQI T1-21 T1-22 T1-23	115 140 185 175 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 500MH .1 MH to 500MH .1 MH to 500MH 10MH to 100MH QUENCY NDUCTORS 2 0/KC TO 10MC .010MH to .150MH .010MH to .700MH		Meets Specifi No Tu No Ma	AMETIC CONSTA TRANS Military cations bes oving Parts	CALLY NT VOI SFORME Accure Fast F Fully	SEAL LTAGI ERS. atte Reg Respons Automa	ED E volatio e atic
TI-18 TI-10 TI-10 TI-19 TI-3 TI-3A HIC TORC FREQI TI-21 TI-22 TI-22 TI-22 TI-22	115 140 185 175 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 500MH .1 MH to 500MH .1 MH to 100MH 10MH to 100MH 00MH to 100MH 20KC TO 10MC .010MH to .150MH .010MH to .500MH .010MH to .50MH		Meets Specifi No Tu No Ma	A military cations bes oving Parts	CALLY NT VOI SFORME Fast F Fully	SEAL LTAGI ERS. atte Reg Respons Automa	ED E oulation e atic
T1-18 T1-8 T1-10 T1-10 T1-9 T1-3 T1-3 T1-3A HIC TORC FREQ T1-21 T1-22 T1-23 T1-20	115 140 185 175 100 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 500MH .1 MH to 500MH .1 MH to 500MH .1 MH to 100MH 0 00MH to 100MH QUENCY NDUCTORS : 20KC TO 10MC .010MH to .150MH .010MH to .500MH .050MH to 5MH		Meets Specifi No Tu No Ma	A military cations bes bes	CALLY NT VOI SFORME Fast F Fully	SEAL LTAGI ERS. ate Reg Respons Automo	ED E oulation e atic
TI-18 TI-8 TI-10 TI-10 TI-9 TI-3 TI-3A HIC TORC FREQU TI-21 TI-22 TI-23 TI-20	115 140 185 175 100 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 200MH .1 MH to 500MH .1 MH to 50MH 10MH to 100MH 000MH to 100MH CUENCY NDUCTORS : 20KC TO 10MC .010MH to .150MH .010MH to .500MH .050MH to 5MH		Meets Specifi No Tu No Ma	A military cations bes bes by ing Parts	CALLY NT VOI SFORME Accure Fast F Fully	SEAL LTAGI ERS. ate Reg Respons Automo	ED E oulatio e atic
TI-18 TI-10 TI-10 TI-9 TI-3 TI-3A	115 140 185 175 100 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305	.1 MH to 100MH .1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 500MH .1 MH to 500MH .1 MH to 500MH .1 MH to 500MH .1 MH to 100MH .1 MH to 100MH .1 MH to 100MH .1 0 MH to 100MH .1 0 MH to 100MH .0 0 MH to .1 50MH .0 10 MH to .1 50MH .0 10 MH to .500MH .0 10 MH to .500MH .0 30 MH to .500MH .0 50 MH to .500MH .0 10 MH to .500MH		Meets Specifi No Tu No Ma	A method	CALLY NT VOI SFORME Accure Fast F Fully	SEAL LTAGI ERS. aute Reg Respons Automa	ED E oulation e atic
T1-18 T1-8 T1-10 T1-10 T1-9 T1-3 T1-3A HIC TORC FREQU T1-21 T1-22 T1-22 T1-20 FREED	115 140 185 175 100 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH 1 MH to 200MH .1 MH to 500MH .1 MH to 500MH 10MH to 100MH 000MH to 100MH .20KC TO 10MC .010MH to .150MH .010MH to .500MH .050MH to .500MH .050MH to .50MH		Meets Specifi No Tu No Ma	A method of the second	CALLY NT VOI SFORME Accure Fast F Fully	SEAL LTAGI ERS. ate Reg Respons Automo	ED For a state E state
TI-18 TI-10 TI-10 TI-19 TI-3 TI-3A HIC TORC FREQ TI-21 TI-22 TI-22 TI-23 TI-20 FREED	115 140 185 175 260 310 GH FRE DIDAL I UENCY RANGE 205 250 210 305 QUALITY SION LAR	.1 MH to 100MH .1 MH to 100MH 1 MH to 200MH .1 MH to 500MH .1 MH to 500MH .1 MH to 500MH 10MH to 100MH 00H to 100MH 20KC TO 10MC .010MH to .150MH .010MH to .500MH .010MH to .500MH .050MH to 5MH	NG	Meets Specifi No Tu No Ma	A Military cations bes oving Parts	CALLY NT VOI SFORME Accure Fast F Fully Com	SEAL LTAGI ERS. ate Reg Respons Automo	ED ulatic e atic
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Туре	Punch through Voltage max.	f _{αb} ave. Mc	H_{FE1} ave. $I_{C} = 50 \text{ mA}$ $V_{CE} = -0.35v$	H _{FE2} ave. (1) I _C = 200 mA (2) I _C = 400 mA $V_{CE} = -0.50v$	I _{co} at -12v μA	r_{b}' $I_{c} = -1 mA$ ohms	C_{ob} $V_{CB} = -6v$ $\mu\mu f$
2N658	-24	5	50	 (1) 45 (1) 65 (2) 75 (2) 100 (1) 65 	2.5	60	12
2N659	-20	10	70		2.5	65	12
2N660	-16	15	90		2.5	70	12
2N661	-12	20	120		2.5	75	12
2N662	-16	8	30 min.		2.5	65	12

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Dissipation Coefficients: In air 0.35°C/mW; Infinite Sink 0.18°C/mW

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(Continued on page 374A)



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Quality

Electronic Components

(Continued from page 368A)

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Stevens-Arnold Modulators

DC-AC Choppers, Ultra-High Speed Relays, Frequency Sensitive Relays, Vibrating Reed Capacitance Modulators, Pictured Above: Twin Contact DC-AC Chopper & Low Drift Vibrating Reed Capacitance Modulator, *DC Drive Chop-



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Stratton & Co. Ltd., Booth 1820 See: British Radio Electronics, Ltd.

Stromberg Time Corp., Booth 1726A See: General Time Corn.

R. H. Sturdy Co., Inc., Booth 1627 See: C & K Components, Inc.

Superior Electric Co. 83 Laurel St. Bristol, Conn. Booths 2722-2732

J. S. Louden, E. S. Williams, R. L. Wiggs, B. G. Deming, I. F. Wolk, P. R. James, K. E. Lang, H. W. Lorenson, R. J. Caccavelli, R. M. Mosher, A. G. Muiler, R. E. Spencer, M. D. Stokem



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Nathan Berro Electronic Equipment Engineering, the only monthly magazine devoted exclusively to electronic research and design offers special Product Di-rectory Section listing exhibitors by product with new products described and keyed with inquiry numbers, . . . Miniaturization Award Winners (first four) will be on display.

(Continued on page 37221)

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Indicates new product.



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These units are used in signal generators, wide-band amplifiers, pulse generators, field intensity meters, micro-wave relay systems, and repeater stations. They find application as laboratory standards, test equipment, and for checking out all types of instruments.

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RFA & RFB 541	10, 20, 20, 20 Db	70		
RFA & RFB 542	2, 4, 6, 8 Db	20		
RFA & RFB 543	20, 20, 20, 20 Db	80		
RFA & RFB 550	1, 2, 3, 4, 10 Db	20		
RFA & RFB 551	10, 10, 20, 20, 20 Db	80		
RFA & RFB 552	2, 4, 6, 8, 20 Db	40		

Other Db loss combinations are available.



Write for complete information

(Continued from page 364A)



Spectra Electronics Corp., Booths 2241-2243

See: Douglas Microwave Co., Inc.

Spectrol Electronics Corp., Subsid. Carrier Corp., Booths 1907-1909 1704 South Del Mar Ave. San Gabriel, Calif.

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Lecture Halls in the Coliseum are located on the Fourth Floor. See complete program of speakers and papers in the editorial section of this issue.

Sperry Microwave Electronics Co. **Division Sperry Rand Corp.** Clearwater, Fla. Booth 2438 E. C. Best, J. Newitt, ▲ G. Eckert, F. Lavelle, ▲ D. Wells, ▲ P. Ely, P. Thomas, J. Duffy, ▲ R. Greenwood, ▲ B. Duncan, D. Mirabella



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Transistors, SBT, MAT, MADT; paper, paper-film, tantalum and aluminum electrolytic; ce-ramic and mica capacitors; magnet wire; pack-aged components; pulse transformers, pulse net-works; interference filters; toroids; magnetic shift registers; wire-wound and film resistors; com puter subassemblies and logic circuits; solid electrolyte dry batteries. See also: Dynacor, Inc., Booth 4020

Stability Capacitors Ltd. (SRC), Booth 1820

See: British Radio Electronus, 1 td.

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Standard Pressed Steel Co., Booth 4127 Jenkintown, Pa.

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Steel Co., Herman D., Booth 4034 See: Swiss Jewel Co.

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(Continued on page 370A)

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using ceramic tubes.

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T 102	1.5	4.5	55	300	- 55 to 100°C		
T 103	3.5	4.5	55	300	— 55 to 100°C		
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SP

(Continued from page 360A)

Sola Electric Co. Div. of Basic Products Corp. 4633 West 16th St. Chicago 50, Ill.

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Sonobond Corp., Booth 4238 See: Aeroprojects, Inc. Sonotone Corporation Electronics Application Div. & Battery Div. Elmsford, N.Y. Booth 1901

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Sony Corporation 514 Broadway New York 12, N.Y. Booth 2004

▲ Kazuo Iwama, Dr. Saburo Uemura, Keiichi Nakamura



*ESAKI "Tunnel Diodes" Transistors, Diodes, *ESAKI "Tunnel Diodes," Other Semiconductor Products.





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FIRST AID ROOM. First mezzanine at north side of floor. Take elevator 20.

LIST OF REGISTRANTS. A complete list of all persons who have registered, brought up to date twice daily, is on the first mezxanine at the back of the first floor. Sorensen & Co., Inc. A Subsidiary of Raytheon Company Richards Ave. South Norwalk, Conn. Booths 2604-2606 L. L. Helterline, Jr., P. J. Deery, R. E. Slater, J. M. Polis, C. B. Woram, B. Campbell



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(Continued on page 368A)

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These examples cover four ranges of Levinthal klystron power amplifiers. All of them are of top commercial quality and all are suited to the requirements of powerincrease problems in existing systems. They can be supplied for c-w or, with various types of modulators, for pulse operation. Further, any of them can be sup-

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This equipment, illustrated above, is typically capable of 50-kw c-w output with 45-db gain in the 300 to 500-mc range, using an Eimac 4KM170,000LA klystron. Pulse capability up to 200 kw can be provided.



3003 and 3004.

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One of these may be the solution for one of your current problems. If not, tell us what your needs are and let us supply information on other examples of our equipment engineering.



2 KW — MODEL 208T

Here, a series of typical klystrons gives a selection of operations with gains from 30 to 50 db over frequency bands including extremes of 385 and 8500 mc.

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Levinthal specializes in the production of high-power r-f transmitters and modulators for c-w or pulsed application. There are presently numerous attractive apenings in the exponding engineering force at Palo Alto. If you are thinking about a new direction in your career, send us your resume and let us tell you something about the general advantages of working on the San Francisco Peninsula and those applying to Levinthal in particular.



LEVINTHAL ELECTRONIC PRODUCTS



RADIATION NCORPORATED

(Continued from page 357A)

Sigma Instruments, Inc. 170 Pearl St. S. Braintree 85, Mass. Booths 2733-2735 P. Garnick, R. B. Wolf, C. E. Heller, H. W. Fleming, W. H. Holcombe, F. C. Burridge, R. H. Pierçe, L. D. LaFlamme, L. B. Stein, J.



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*New Transistor Guard, Model 101

Communications Antennas, Filters, Duplexers, Band-Pass Cavitics, Antenna Decouplers, Repeater Duplexers, Common Antenna Multicomplers, Antenna Test Set, Transistor Guard-Solid State Circuit Breaker.

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> Skydyne, Inc. River Rd. Port Jervis, N.Y. Booths 4515-4515A

▲ R. L. Weill, W. F. MacCallum, R. D. Cooper, G. B. Parsons, Steven E. Mautner, Lawrence A. McFadden, Victor Orben, W. M. Clevenstine, E. M. Porter



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Ira L. Landis, Sheldon Sackman, Charles Gold, Nat Kopf, Harry Weston



New Products of Herman H. Smith, Inc.

Phone Tip and Banana Plugs, Jacks, Binding Posts, Alligator Clips, Test Leads, Patch Cords, Phone Plugs and Jacks, Panel Indicators, Knobs, Turret Terminals, Shaft Accessories, Terminal Lugs, Strips, Boards, Hi-Fi Cords, Sockets, Hardware, Switches, New (illustrated): Dual Banana Plugs, Junibo Clips, Pilot Light Shields, Nylon Banana Jacks.

Sodeco-Geneva, Switzerland, Booth 3012

See: Landis & Gyr, Inc.

(Continued on page 364A)

First and Second floors—Components Third floor—Instruments and Complete Equipment

Fourth floor-Materials, Services, Machinery

Developments at the IRE Show



ON THE SECOND FLOOR, to the right of the escalator, you'll find this Raytheon exhibit of electronic component and equipment advances-N. Y. Coliseum, Booths 2604-2614. SEMICONDUCTOR DIVISION

New "Avalanche Mode" Silicon Transistor switches in $2\frac{1}{2}$ milli- μ sec.

A guaranteed switching time of 10 millimicroseconds maximum (when used in the switching circuit shown) with speeds faster than $1\frac{1}{2}$ millimicroseconds in some applications is now possible with Raytheon's new



2N1468 Silicon NPN transistor for avalanche mode operations. Other features: 40 watts peak power, average power dissipation of 250 milliwatts, maximum operating temperature of 125°C.

New Silicon Mesa Transistor in Subminiature Package.

Raytheon's Semiconductor Division announces the availability of highperformance Silicon Mesa transistors in subminiature packages (.130" D., .160" H.). These units feature the reliable "Mesa" construction, alpha cut-off frequencies up to 50 megacycles and close control of DC base-current gain in high-speed switching types.



MICROWAVE AND POWER TUBE DIVISION

Four new ruggedized backward wave oscillators cover 1-12.4 kmc.

Four compatible Raytheon BWO's now provide continuous frequency coverage from 1 to 12.4 kmc. They utilize interdigital delay-line structures for greater ruggedness and heat dissipating characteristics, are smaller and lighter than their predecessors and have improved finegrain tuning variations with minimized fine-grain power output variations. Forced air cooling is not required under normal operating conditions.



New ferrite circulators for masers, parametric amplifiers and radio astronomy. A standard line of extremely compact, low-frequency, three-port ferrite circulators is now available from Raytheon. Now, a total of



able – for UHF as well as Sand L-band applications. The new UHF unit extends the frequency range down to 400 mc. These circulators are supplied with fixed permanent magnet fields (as illustrated) or with tuned magnetic fields for full performance over a broader band.

MACHLETT LABORATORIES

UHF planar triode has 60% more cathode current capacity. This unique.

Machlett developed UHF planar triode with ceramic envelope has 1.6 times the cathode current rating of the more conventional tubes in current use. The new ML-7211 has applications in communications, navigation, telemetering, radar and missile equipment of the most advanced design.



Improved camera tube for general closed-circuit TV applications. Machlett's novel vidicon camera tube features photoconductor guard and self-aligning beam eliminating

the need for permanent magnets or coils. This tube will be exhibited at the IRE along with Machlett's watercooled triode rated at 20-megawatts peak power, a scan conversion tube for conversion of radar information to TV display, and a highpower vapor cooled tube for general purpose modulator. amplifier and oscillator services.



RAYTHEON COMPANY Waltham, Massachusetts



Excellence in Electronics

.

See these new Raytheon Product



INDUSTRIAL COMPONENTS DIVISION

New Kilo-line recording storage tubes provide 1000 TV lines at 50% modulation. These highresolution, low-noise tubes for frequency and scan conversion utilize a



specially designed tetrode electron gun for a resolution of 1,000 TV lines at 50% modulation and provide better control over beam cut-off than conventional triode guns. Applications include: (1) scan conversion for bright display radar and moving target indicators, (2) slowdown video for still picture telephone transmission, (3) stop-motion video.

New pointer knobs virtually eliminate parallax. Two new sloping pointer style control knobs have been added to Raytheon's widely used commercial and military knob line. The new pointer series complies fully with military specifications and is available in black or



grey, with or without dial skirts; in mirror or nonreflective matte finish. Colors are available on special order. Raytheon's complete line of knobs includes 206 styles -9 standard types in 6 sizes, plus tactile shapes, color and color caps. COMMERCIAL APPARATUS & SYSTEMS DIVISION

New Voltage-regulating PF transformer holds voltages to within \pm 3%. The new Raytheon voltage regu-

lating PF transformer maintains plate and filament voltages to within $\pm 3\%$ of rated output with line voltage variation of from 100 to 130 volts. PF transformers are now available in three standard models with ratings up to 380 VDC at 250 MA. They eliminate need for VR tubes and special circuitry.



Sorensen Series Q line of power supplies for 6, 12 or 28 VDC regulated. The Sorensen Series Q power supply line is comprised of 15 different models with outputs of 6, 12 or 28 VDC, adjustable approximately \pm 25%. Voltages are regulated within



 \pm 0.05% for load and line combined. Power capacities range up to 200 watts. The complete line of Sorensen power supplies covers requirements from 600,000 volts down to 3 volts. Sorensen also offers a line of frequency changers and line voltage regulators.

Visit us at the New York Coliseum, Booths 2604-2614

Shockley Transistor Corp. Sub. Beckman Instruments, Inc. Stanford Industrial Park Palo Alto, Calif.

Booth 1205

H. S. Schuler

Four-layer silicon diodes (for switching), Compensated Avalanche Diodes (for voltage regulation).

Shurite Meters, Booth 2833 See: J-B-T Instruments, Inc.

Sibley Company, Booth 1100B Bridge St. Haddam, Conn.

W. F. Moore, D. Dewey, M. Whiston, B. Roffman, R. Murray, W. Murray, J. Churchill, R. S. Pettigrew, H. Carvey

Printed circuits, flush commutators, drum commutators, plated-through hole circuits. Precious metal plating of electronic parts. Electronic assemblies. Engineering research and development, department for circuitry conversion, miniaturization and high temperature applications.

F. W. Sickles Co., Booths 1218-1224 See: General Instrument Corp.

Siegler Corp., Booths 1427A-1427B & 3214

See: Magnetic Amplifiers, Inc. & Bogen-Presto Co.

Sierra Electronic Corporation, Subsid. Philco Corporation, Booths 3031-3032 3885 Bohannon Drive Menlo Park, Calif.

C. M. Volkland, H. D. Farnsworth, M. J. Gothberg, S. K. Ashby, ▲ W. Feldscher, ▲ S. Frankel, ▲ G. K. Patterson

* 5. Frankei, & G. K. Fatterson * Frequency Selective Volumeters, *Calorimeter Systems, *Harmonic Filters, *FM Signal Generator (SL band), Line Fault Analyzers, Oscilloscopes, Transistor Tester, Power Sources, Power Monitors, Termination Wattmeters, Directional Couplers, Stub Tuners, Coaxial loads, Waveguide Loads, Ion Gauges.

Sierra Research Corporation, Booth 3015

P.O. Box 22

Buffalo 25, N. Y.

John P. Chisholm, Harold K. Fletcher, ▲ Herbert Mennen, ▲ Vernon H. Siegel, Robert J. Theisen, ▲ Sherwood H. Calhoun

Product History: Research, development, and fabrication of data transmission equipment: 'PAM Coders and Decoders; computer accessories: Analog Multipliers; radar equipment; 'Broadside Array Antennas, "Transistorized Indicators, "Target Simulators; and ship motion recording instrumentation: "High-level Torque Meters.

(Continued on page 360A)

▲ Indicates TRE member. 7 Indicates new product.

See all the exhibits!

Don't miss these important locations—

Mezzanine at back of first floor, South Room at center of south wall, second floor. 3000 court at southeast corner of third floor, 4000 court at southeast corner of fourth floor, 4500 court in northwest corner of fourth floor.





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Whom and What to See at the Radio Engineering Show

(Continued from Juge 352.4)

Servo Corporation of America, Booths 3806-3808

111 New South Road Hicksville, L. I., N. Y. Walter Campbell, Arthur Freed

Electronic Equipment, Pyrometer, Servo System Test and Analysis Equipment and Systems, Model Servotherm®, Servoscope®. See also: Electro-Pulse, Inc., Booths 3810-3812

Servo Dynamics Corp., Booths 1401-1407 Subsid. National Co., Inc. Somersworth, New Hampshire

R. H. Rogers

Servo motors, damping generators, tachometers, gear heads, inertia damping servo motors, servo subsystems.

Servomechanisms, Inc., Mechatrol Division, Booth 2812

1200 Prospect Ave.

Westbury, L. I., N. Y.

Victor See, E. J. Chevins, L. R. Pensiero, R. N. Sebris, N. A. Christian, G. L. Dinger, D. Ladner

D. Ladner Serve Motors, Hysteresis Synchronous Motors, Integral Gearhead Motors, Viscous and Inerti-ally Damped Servo Motors, Damping and Tem-perature Compensated Integrating Tachometers, Stepper Motors, Clutches, Electronic and Me-chanical Breadboard Parts Including Amplificrs, Power Supplies, Modulators (60 and 400 cycles), Counters and Potentiometers.

Shallcross Mfg. Co., Booth 2634 Selma, N. C.

John S. Shallcross, Don M. O'Halloran, Dewees H. Shallcross, Jr., R. I. S. Crisp, R. A. Avery, Robert W. Mills, Clayton Huber

Precision Resistors, Rotary Switches, Instru-ments, Delay Lines, Resistor Networks, Audio Attenuators, Coils.

Shepherd Industries, Inc. 103 Park Ave. Nutley 10, N.J. **Booth 2937**

John French, Dan Giffin, A Dr. William Duerig, A William Murphy, Art Leu-thesser, A Don Killen, L. L. Driggs, Dan Coll, Dave Carpenter

Digital Tape Transports, *Bin Type and *Airborne; *Digital Magnetic Memory Drums; *Magnetic Heads for Tape Trans-ports and Drums; *Logic Circuitry, *Solid State Amplificrs; *Analog Tape Transports.

Shielding, Inc. 514 North Read Ave., P.O. Box 3 Riverton, N.J. Booths 3061-3062

▲ J. W. McDonald, Jr., ▲ T. P. Reath, J. J. Mooney, J. J. McDevitt, W. J. Ryan, D. J. Shamp



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- ... because there can be no loosening of the seal due to compression set.



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(Continued from page 352A)

Security Devices Laboratory Electronic Div. Sargent & Greenleaf, Inc. 24 Seneca Ave. Rochester 21. N.Y. Booth M-22 Joe Williams, Ron Fisher, Art Scalzo



Resonant Reed Relay; Oscillator Control

Reson-ator J610 Oscillator and J500 Relay Controls are extremely stable electro-mechanical component devices used to generate and decode specific alternating current signals. Also see our BF series Battery Holders to Accommodate all Batteries or Cells used in Subminiature Equipment.

Semiconductor Products, Booth 4215 See: Cowan Publishing Corp.

> Be sure to see all four floors!

Sensitive Research Instrument Corp. 310-316 Main St. New Rochelle, N.Y. Booths 3410-3412

Marvin I. Steinberg, Leonard J. Patterson, ▲ H. Russell Brownell, Louis Miller, Robert Most, F. Patrick Johnston, Mike Kane, Gerald D. Frank, Earl Elliott, Abe Isaacs



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Servel Inc., Booth 2806 See: Burgess Battery Co.

(Continued on page 356A)



CONSTRUCTION CO. 2733 HAWKEYE DRIVE SIOUX CITY, IOWA



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Tell-tale monitoring — all controls easily accessible, up front, Tiny tell-tales spot trouble instantly, isolate it to a specific modular unit,



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— a!l Single point of check leads brought to single checkpoint — makes possible color coded, vividly illus-trated circuit legends,



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3



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PRODUCTS COMPANY, 3121 N, Main Street, Bockton, Mass, See you at the IRE Show—Booths 1508, 1510

PROCEEDINGS OF THE IRE

March, 1960



(Continued from page 349A)

Sanborn Co. 175 Wyman St. Waltham 54, Mass. Booths 3601-3605 Steven Bilowich, Ralph E. Hanson, & R. Paul Foster, E. D. Pulsifer, Alfred E. Lonnberg, A Dor. Arthur Miller, & J. William Sauber, Morton H. Levin, & Robert T. Danehy, Donald M. Brown



Model 320 Dual Channel Recorder

On display are these new DIRECT-WRITING recorders: rack-mounted 2-channel, portable 2channel, rack-mounted 16-channel, optical 24channel. New 8-channel 17-inch oscilloscope. Also on exhibit are: 350, 850, 950 Series equipment; Model 670, A. NY Recorder; pressure, displacement, velocity transducers.

Sanders Associates Inc. 95 Canal St. Nashua, N.H. Booth 1723

▲ Don Ayer, John Killelea, Harvey Hollister, Tom Dolan, Al Meyers, Lou Garten, Ken Glover, Morris Silverman Modular components for microwave strip transmission including new crystal mount for MICRO-MIN dudes; Subminiature rate gyroscopes; FLEXPRINT flexible printed wiring for jumpers, cables, harnesses; miniature blowers.

Sanford Mfg. Corp., Booth 4038 See: Micromech Mfg. Corp.

Sangamo Electric Co., Booth 1921 11th and Converse Sts. Springfield, Ill. William S. Paine

Capacitive and Inductive Components, Mica, Paper, and Electrolytic Capacitors, Specialty Filters, Transformers, Pulse Networks and Toroidal Coils.

Sargent & Greenleaf, Inc., Booth M-22 See: Security Devices Laboratory

Saxton Products, Inc., Booth 4055 4320-26 Park Ave. New York 57, N.Y.

Edward Abbo, Mark Kleiner, Jerome Firestone, Conrad Deutson

Wire & Cable Div. of Saxton Products maintains large stock CO-O Cable (Flexible & Extra Flexible 300 & 600 volts) MIL-C-34322A, CO-S types MIL-C-3884, also Teflon Types "E" and "EE" MIL-C16878C. Specialists in Cable To Your Specification Including Short-Runs. Schutter Microwave Corp., Booth 1726B 80 East Montauk Highway Lindenhurst, N.Y.

▲ Norman Kjeldsen, ▲ Milton D. Hirsch, ▲ Edward Buckley, ▲ Dan McDonald, ▲ Frank Lippman

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Scientific-Atlanta, Inc. 2162 Piedmont Rd., N.E. Atlanta 9, Ga. Booth 3909

▲ Wm. H. Bradley, ▲ Glen P. Robinson, Jr., ▲ J. Searcy Hollis, ▲ Herbert W. Bass, Herbert Gentry, M. J. (Bud) Mc-Donald, ▲ J. Emory Lane

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Sealectro Corp., Booth 2922 139 Hoyt St.

Mamaroneck, N. Y. George E. Mohr, Milan E. Robich, Robert O. Walcovy, William Silberstein, Augustus S. True, Larry Willis, George Bechtold, Vincent E. Malkiewicz

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Secon Metals Corp. 7 Intervale St. White Plains, N.Y. Booth 4117 ▲ Eugene Cohn, Richard Gordon



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(Continued on page 352.4)

▲ Indicates IRE member.

* Indicates new product.

The letter "M" preceding a booth number indicates that the exhibitor will be found on the mezzanine at the back of the first floor.

Sage Electronics Corp. **Country Club Road** Rochester, N.Y. Booth 2236 F. Dwight Sage, Davidge H. Rowland, Allen P. Mills, ▲ J. C. Van Arsdell



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Sage Laboratories, Inc., Booth 1910 3 Huron Drive

East Natick Industrial Park Natick, Mass.

G. A. Ayoub, J. A. Camuso, ▲ J. J. Chacran, W. J. Kennedy, E. W. Lattanzi, ▲ T. S. Saad, J. M. Shalhoub, ▲ R. Tenenholtz Stripline Hybrids, *Coaxial Directional Coup-lers, Cobrids, 'Balanced Mixers, Rotary Joints, Slotted Lines, Crystal Holders, Isolators.

Howard W. Sams & Co., Inc., Booth 4048

2201 East 46th St.

Indianapolis 5, Ind.

J. A. Milling, W. D. Renner, B. C. Landis, L. H. Nelson, G. Mowry, T. G. Shonfield. Specialized Services and instruction manuals.

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San Diego Scientific Corp., Booth 3930B 3434 Midway Drive

San Diego 10, Calif.

▲ David C. Kalbfell, John W. Bodnar, John J. Chaparro, ▲ Philip Wasserman

Chaparro, A Philip Wasserman Magne-Plexer®, low level solid state commutator for millivolt signals. Operates at 6000 samples per second, available practically any number of in-puts. Floating inputs provide common mode re-jection in order of 135db. Output 5 volts, typical accuracy of 0.2%. Signal circuits not switched; programming readily changed by plug-in cards.

San Fernando Electric Mfg. Co., Booth 2635

1509 First Street

San Fernando, Calif.

Alan Rubendall, Michael Rosenberg, Jerry Ciral, Lyle R. Smith, Donald E. Rubendall, Eileen Johnson, Helen Jensen, Gene Bell, Ker-mit Hawkins, William Gutknecht

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(Continued on page 350.4)

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Third floor—Instruments and Complete Equipment

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Whom and What to See at the Radio **Engineering Show**

(C ntinued from page 3464)

Rotating Components, Inc., Booth 1234 267 Green St.

Brooklyn 22, N.Y N. F. Giarra, A. G. Miller, J. H. Mills, G. Sprinsky

Sprinsky Inverted Blowers—cool running, compact, long life, high pressure, Mil spec ac motors, blowers, axial fans, tube & vane axial fans; cooling cabinets, servo motors, gearmotors, generators, 1/2000 to $^{1}_{4}$ h.p., 1" to 35_{16}^{**} " OD, low & high temp, variable & single freq. 50 to 1000 cps. Stock or custom.

Rotron Mfg. Co. 7-9 Schoonmaker Lane Woodstock, N.Y. Booths 2830-2832

▲ J. C. van Rijn, P. M. Beard, J. C. Larson, D. Carlson, W. W. Blelock, Jr., F. Desmond, H. E. Huber, J. P. Cerasaro, P. S. Lyon, A. Gran, L. Sima, E. N. Goddard, M. Edwards, G. M. Taylor



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Rubicon Div., Booth 2210 See: Minneapolis-Honeywell.

Rutherford Electronics Co., Booth 3834 8944 Lindblade St. Culver City, Calif.

▲ C. E. Rutherford, ▲ Howard E. Mette, Stan Schwartz, ▲ N. T. Holzer

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Rye Sound Corp., Booth 2933 See: Richard Hirschmann Radiotecnisches Werk

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(Continued from page 344A)

Reon Resistor Corp. 155 Saw Mill River Rd. Yonkers, N.Y. Booth 1926 ▲ Leon Resnicow, J. J. McCann, Eugene V. Mandel, Mortimer Lazarus, Eric Maneskjold



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Rex Corp., Booths 4306-4310 See: William Brand-Rex Div.

Rheem Semiconductor Corp. 327 Moffett Blvd. Mountain View, Calif. Booths 2925-2926

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Richards Electrocraft, Inc., Booth 2126

See: G-C Electronics Mfg. Co.

Riggs Nucleonics, Booth 3018 See: ELIN Division.

Rixon Electronics, Inc. 2414 Reedie Drive Silver Spring, Md. Booth 3411 ▲ J. L. Hollis, ▲ C. J. Harrison, ▲ J. C. My-rick, D. W. Perry, M. F. Frank



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(Continued on page 348A)



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March, 1960

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(Continued from page 343A)

Raytheon Company, Commercial Ap-paratus & Systems Division, Magnetic Operations, Booths 2610-2614 190 Willow St.

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Raytheon Company, Sorensen & Co., Inc., Booths 2604-2606

See: Sorensen & Co., Inc., Subsid. Raytheon Co.

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(Continued on page 346A)

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▲ F. A. Gunther, J. R. Day, H. Saden-water, M. Kraus, ▲ G. Papamarcos, T. P. Rizzuti, ▲ H. H. Robinson, D. F. Koi-jane, A. Waxman scatter radio euniument.

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See: Thompson Ramo Wooldridge, Inc.

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Raybestos-Manhattan, Inc., Asbestos Textile Div., Booth 4014 Manheim, Pa.

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(Continued on page 344A)



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551	NR	5	KVDC	5	21/2	x	31/	1 3 1/	3.	25	÷.
1051	NR	10	KVOC	5	3 1/4	н	4 %	1 61/2	8#	10	οz.
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WRITE FOR LITERATURE



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Whom and What to See at the Radio Engineering Show

(Continued from page 340A)

Pyle-National Co., Booth M-4 1334 North Kostner Ave. Chicago 51, Ill.

▲ H. F. Whalen, Jr., J. F. Shearer, L. J. Milewicz

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See: Magnetic Instruments Co., Inc.

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See: Technical Devices Co.

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Radio Corporation of America, Booths 1602-1608, 1701-1707 Defense Electronic Products

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(Continued from page 338A)

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(Continued on page 342A)



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944 Long Beach Avenue, Los Angeles 21, Calif.
Whom and What to See at the Radio Engineering Show

(Continued from page 336A)

Philco Corp. Lansdale Tube Co. Div. Church Road Lansdale, Pa. Booths 1302-1308 W. J. Peltz, ▲C. H. Warshaw, W.
F. Maher, M. J. James, ▲S. L. Levy,
▲C. S. Simmons, ▲C. I. Swanson, K. E.
Schubert, ▲E. S. Eisenscher, J. J.
McCartin, J. Kindregan, W. Brydia,
D. E. Reynolds, P. S. Armstrong, E. W.
Bobigan, A. Berry, M. L. Snyder, J. W. Mintzer Minizer Philco's high-power, high-speed Micro-Alloy Diffused Base (MADT*) transis-tors, complete transistor line. Millimeter Diode* and others, IR Detectors. Pano-rama of complete communications sys-tem with semiconductors; radar, ground control, computers, missile tracking gear. Working 'scope shows H-I' MADT switch performance in circuit.

Philco Corp., Booths 3031-3032 See: Sierra Electronic Corporation.

Philips Electronics, Inc., Booths 2522-2530

See: Amperex Electronic Corp., Electrical In-dustries, Ferroxcube Corp. of America

Phillips Control Corp., Booth 2340 59 W. Washington St. Joliet, Ill.

Merle Hayward, W. R. Baughman, Ralph Nie-land, J. Rowell, ▲ H. H. Hartong, Bill Faci-nelli, ▲ Bill Reagan, Don Schofield, Harry Buys, Dick Smego

Quality Relays for—Aircraft; Multi-contacts featuring longer life; Power & General Pur-pose; Solenoids; See new Crystal Can & Versa-tile Telephone Types; Hermetic Seals of All Types; Check unique Phillips Dependability in products, engineering & Service!

Photocircuits Corp., Booths 2201-2203 31 Sea Cliff Ave.

Glen Cove, L.I., N.Y.

H. I. Rudman, & J. Calpena, & F. Beste, & R. L. Coryell, & S. Hudson, & A. W. Kelly, & J. M. Knizeski, & S. H. McVicar, & P. L. Ross, & C. J. Saffery

Conventional and miniaturized printed wiring boards utilizing Tuf-Plate plated-through hole design. Proto Division offers fast delivery serv-ice for prototype and sample quantities. Printed circuit motors photoelectric tape readers, master circuit design technique, ceramic printed circuit.

Physical Sciences Corp., Booths 3707-3709

See: Packard-Bell Electronics

Pic Design Corp., Booth 1625 477 Atlantic Ave.

East Rockaway, L.I., N.Y. Winfred M. Berg, Rowland F. Schwenker, P. J. Wellenberger, John L. Swane, Charles Kee-nan, Herman Hering, Jack Bradley, William Swane, Walter Caverno

Precision 2 and 3 gears from stock. Precision Breadboard & Development Components: Instru-ment Plates, Bevel Gear Boxes, Magnetic Clutches, dials, Differentials, shafts, Tool Parts, etc. *Just Released anti-backlash worm wheel and Zero Adjustable Bellows Coupling.

MORE DATA

Exhibitors shown in boxed listings, or with product illustrations, have more data for you in their advertisements in the March 1960 issue of "Proceedings of the IRE." Pickard & Burns, Inc., Booth 3044 240 Highland Ave. Needham 94, Mass.

I. J. Metcalfe, ▲ F. C. Leiner, ▲ M. N. Ar-lin, ▲ M. F. Spears, ▲ J. J. Glynn, ▲ T. C. Cahill, S. L. Smith, ▲ L. D. Shapiro, J. C. Wil-liams

Research. Development and Manufacturing in the fields of Radio Communications and Naviga-tion Systems, Instrumentation and Antenna Sys-tems. A new Radio Locked Frequency Stand-ard (RALOC); Temperature Monitoring Equip-ment

Pitometer Log Corp., Booth 3024 237 Lafayette St. New York 12, N.Y.

Bernard Kahn, D. J. Kreines, & Norman Sturm, A R. C. Rosaler, Rudy Volhard, Ed Bleistein, A Gordon Silversmith, A Ed Wrobel, Michael Smolin, Bill Hallock

"Series 800 Stalo Tester with L, S and X-Band Tuning Units. Tunable Stalos Covering *UHF, I., *S and X Microwave Frequency Bands, With Stability of 5 Parts in 10⁹. Also Xtal Chain Stable Local Oscillators. High Tempera-ture Transistorized Controls.

Pittman Electrical Developments Co., Booth M-21 Sellersville, Pa.

Charles A. Pittman III, William G. Blacklock

Miniature low voltage permanent magnet field direct current motors.

Plastic Capacitors, Inc. 2620 N. Clybourn Ave. Chicago 14, Ill. **Booth 2740**

Stephen Meskan, Harold Francis, Robert Cur-tin, Richard Henry, Richard A. Strassner, Mer-rill Holt, Dave Dolin



*New close tolerance polystyrene, *high reliability paper-mylar and *high package density metallized mylar capacitors. Mylar, polyethelene, tefton, and paper dielectrics exceeding MIL requirements. Pulse Forming Networks, 60 and 400 CPS input low current, high voltage power supplies, *low cost impedance comparison bridges.

Be sure to see αll four floors for a complete view of 800 new ideas! **Plastoid Corporation** 42-61 24th St. Long Island City 1, N.Y. Booth 4525

W. Grant, D. J. Nichols, Dean Haggerty, Mil-ton Weinschel, A. W. Anderson, A Leon J. Brodsky, Warren Moffett, C. Myslinski, E. H. Cooper, A Jerry Tomey, T. E. Gaess, Toby Del Guidice

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Tefion Coax Cables

Specifications of wire & cable constructions for hook-up, power communication and amtrol, air-craft, coaxial cable, seismograph cable, gun con-trol, atomic energy, score board, switchboard, underwater sound, airport control, electronic in-struments, power cable per IPCEA Spec, kite cable, shipboard cable, special constructions.

Polarad Electronics Corp. 43-20 34th St. Long Island City 1, N.Y. Booths 3205-3211

A. A. Goldberg, R. J. Sheloff, R. Savold, R. Saul, P. H. Odessey



Universal Spectrum Analyzer

Model SA-84W Universal Spectrum Analyzer Model SA-84W Universal Spectrum Analyzer capable of precisely measuring all microwave parameters; Model RW-T Antenna Pattern Microwave Receiver—frequency range 2 to 75 kmc with single tuning head. Complete line microwave signal generators, spectrum analyzers and receivers. Information available on Pola-rad's Mobile Field Laboratories and Demonstrator.

Polyphase Instrument Co., Booth 2839 East Fourth St.

Bridgeport, Pa.

E. C. Capuzzi, D. J. Seifert, A. M. Meyer, E. Bard, R. F. Adams

Bard, R. F. Adams Pulse, Audio, Toroidal, Converter, Specialty Electronic Transformers, Electrical Wave Fil-ters, Inductors, Delay Lines, Modules* and Special Magnetic Components for Military and Industry. Static and Dynamic Strain Measur-ing Instruments and Accessories. Internally Strain-Gaged Transducers, Load Sensitive Bolts. "New group of standard Lumped Constant Delay Networks.

Polytechnic Research & Development Co., Inc. 202 Tillary St. Brooklyn 1, N.Y.	
Booths 3602-3606	
▲ H. C. Nelson, ▲ M. Wind, ▲ P. Mari- otti, ▲ L. H. Fisher, D. Cooper, W. A. Weissman	
"Pacemaker" line of Microwave & Elec- tronic Test Equipment. *Klystron & BWO/TWT Power Supplies, *Slotted Sections, *Variable & Fixed Attenuators, Sliding Shorts, Standard Mismatches, Sliding Terminations, Rotary Standing Wave Indicators, Bolometers & Thermis- tor Mounts, Frequency Meters, *Stand- ing Wave Amplifiers, Power Bridges, Calorimeters, ctc.	

(Continued on page 340A)

UNIFORMITY!

30

MOTOROLA MESA TRANSISTORS

MOTOROLA 2N700 MESA UHF AMPLIFIER TRANSISTOR

INTOO BYCED DISTRIBUTION AT 25°C

AMPLE OF 150 UNITS

18

20

BVCED (MINIMUM - 25 VOLTS

25

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INC 100 80,

Collector to Base Voltage, Ven Collector to Emitter Voltage, Veg D.C. Collector Current, Ig Maximum Junction Temperature, T Maximum Frequency, fmax

Base Resistance, rb' Common Emitter Current Gain, h_{fe} Collector Capacitance, Cob



Collector Dissipation Pc (25°C Ambient) 75 mw Collector Cutoff Current, I_{CO} typical $V_{CB} = 6V$

HIGH FREQUENCY CHARACTERISTICS





Here is graphic evidence of Motorola Mesa uniformity. Data is typical of that compiled on random samples from all production lots.

The remarkable uniformity of Motorola Mesa transistors means greater reliability...simplified circuit design through better balanced devices ... more reliable circuits. Uniformity also means availability. Because of the unique features of Motorola Mesa transistors, each production line produces only one device ... not a series of similar devices. This enables Motorola to achieve volume production of these sophisticated devices.

FOR TECHNICAL AND APPLICATIONS INFORMATION contact your Motorola Semiconductor district office, today!

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CLIFTON, NEW JERSEY 051 Bloomfield Avenue Regory 2 5300 rom New York WI 7 2980 DETROIT 27. MICHIGAN 13131 Lyndon Avenue BRoadway 3-7171

CHICAGO 39, ILLINOIS 5234 West Diversey Avenue AVenue 2-4300

NNEAPOLIS 27, MINNESOTA 7731 6th Avenue Liberty 5-2198

HOLLYWOOD 28, CALIFORNIA 1741 Ivar Avenue HOllywood 2-0821



PROCEEDINGS OF THE IRE

March, 1960

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Whom and What to See at the **Radio Engineering Show**

Panoramic Radio Products, Inc. 520 South Fulton Ave. Mt. Vernon, N.Y. Booths 3315-3317



Spectrum Analyzers from Subsonic thru micro-wave (0.5 cps to 44000 mc), Response curve tracers from 0.5 cps to 15 mc, FM/FM Telemetry Test Equipment, New 11 point telemetering Cali-brator; simultaneous 18 channel output; all tran-sistorized, 7" high. Improved Single Sideband Analyzer SSB-3a, Broad Band curve tracer, G-6. Unique Vicrowave Analysis SPA-4, 10 mc-44000 mc nique me

Parker-Hannifin Corp., Booth 4243 See: Parker Seal Co.

Parker Seal Co., Booth 4243 10567 Jefferson Blvd. Culver City, Calif.

Frank Opatrny, Don Lee, George Moss, Albert Wickson, Ed Schaub, Jack Watson, Nelson Harway

Standard and special seals for Waveguides and microwave equipment. Seals are reusable, pre-vent R/F leakage, provide no-leakage mechani-cal sealing. Also seals for prototype, short pro-duction runs, and extreme high temperature and low temperature seals for flanges, fittings and features. fasteners.

Par-Metal Products Corp. 36-62 49th St. Long Island City 3, N.Y. Booths 4302-4304 A. A. Parmet, John Novak, James Berry



Four basic types of slide assemblies of our Universal Cabinet Racks. All Welded Universal Cabinet Racks for 19", 24", 30" Wile Panels, with solid side walls or intermediate side panels, Utility Desk Assemblies.

Pascol Division, Booth 1506

See: Licon Switch & Control Div., Illinois Tool Works.

FIRST AID ROOM

A nurse is in charge at all times. First aid room is on the first mezzanine at the north side of the building. Take elevator 20 from any floor.

(Continued from page 334A)



Magnetic Amplifier Regulated D.C. Power Sup-plies; Line Voltage Regulators; "Transistorized D.C. Power Supplies Low Voltage High Cur-rent Type; "Transistorized Inverters & Converters



Static inverters, permanent magnet alter-nators, thermo-electric generators, axial flow fans, electronic cooling packages, en-gine-driven generator, electric motors, secondary power systems, auxiliary power systems

Phaostron Instrument & Electronic Co., Booth 1516 151 Pasadena Ave.

South Pasadena, Calif.

I. W. Eisenberg, H. J. Veitch, Norman Logan Panel meters, laboratory standards, sensitive re-lays, test equipment, airborne components.

Pharmaceutical Industries Corp., Booths 2526-2528 See: Electrical Industries

Phelps Dodge Copper Products Corp.

Inca Manufacturing Division Fort Wayne, Ind. Booths 4028-4029 A. F. VanRanst, H. E. Boe, R. Sutman, J. Matthews, C. Frame, Ralph Hall, D. Hilker, J. Burgoon

The world's most diversified line of magnet wire. The only complete line of solderable magnet wires especially for the electronics industry. And, a new mystery wire of unusual interest to manufacturers of motors, coils, and transformers.

Philamon Labs., Inc. 90 Hopper St. Westbury, L.I., N.Y. Booth 3111 ▲ Boris F. Grib, ▲ Robert A. Hunt, ▲ Jack Miller, ▲ Donal F. Gehring



Tuning Fork Controlled Frequency Package

Tuning Fork Frequency Standards—Primary and Secondary—Transistorized Frequency Gen-erators, Binary and Decade Dividers, Signal Amplifiers, Low Power Amplifiers, Oscillator Circuits for missile, aircraft and ground applications.

(Continued on page 338A)

▲ Indicates IRE member. * Indicates new product.

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

Don't miss this special report at the IRE Show...

NEW DEVELOPMENTS IN HIGH-TEMPERATURE DIELECTRICS

A comprehensive lecture-demonstration by the engineering staff of Mycalex Corporation of America

Time – 3:15 P. M. Daily Place – FRANCE ROOM Floor 2M, N.Y. Coliseum Dates – March 21-24 (Inclusive) Admission – By Invitation ONLY

REGISTER in advance at MYCALEX BOOTH No. 2729-2731



PROCEEDINGS OF THE IRE March, 1960

Just a few of the highlights featured at the Mycalex Corporation of America demonstration will include: applications, engineering problems and solutions in high-temperature dielectrics illustrated with special color slides showing:

- MYCALEX[®] glass-bonded mica
- SUPRAMICA[®] ceramoplastic
- SYNTHAMICA[®] synthetic mica

- complete with materials and component samples. This will be followed by a discussion period during which questions will be answered by the Mycalex engineering staff. For your personal invitation, register in advance at the Mycalex booth No. 2729-2731, 2nd Floor, IRE Show.

General Offices and Plant: 226 Clifton Blvd., Clifton, N. J. Executive Offices: 30 Rockefeller Plaza, New York 20, N. Y.

World's largest manufacturer of glass-bonded mica, ceramoplastic and synthetic mica products

World Radio History

Whom and What to See at the Radio **Engineering Show**

(Continued from page 333A)

Ohmite Manufacturing Co. 3601 Howard St. Skokie, III. Booths 2333-2335 Roy S. Laird, Kenneth M. Arenberg, Edward A. Rehe, Manny Forester



Model "E" 121/2 Watt Miniature Rheostat

*Etched foil tantahum capacitors; 'two new larger sizes plain foil tantahum capacitors; 'ceranic wafer metal film, micromodule resistor; tiny 1 watt wirewound resistor; *7 watt mulded power precision resistor; *20 ampere variable transformer; *new line 30 volt high amperage variable transformers. Rheostats, Switches, Chokes, Diodes.

Oliver-Shepherd Industries, Inc., Booth 2937

See: Shepherd Industries, Inc.

Optical Coating Laboratory, Inc., Booth 4049 977 Sebastopol Rd.

Santa Rosa, Calif.

▲ Rolf Illsley, Danforth Joslyn A koll Hisley, Danforth Josiyn Development and manufacture of vacuum de-posited thin film coatings. Infrared Spectrum: Wide Band Pass, Long Wavelength Pass, Narrow Band Pass Filters, 2-color filters, Visible Spec-trum: Gunsight Reflectors, "Hot" Mirrors, "Cold" Mirrors, High Reflecting Mirrors, Multi-laver Antireflection coatings. Infrared Heat Shielding on Fiberglas.

Optimized Devices, Inc. 864 Franklin Ave. Thornwood, N.Y. Booth 3016

Arthur Zuch, James Murtha, Sergio Bernstein



Caftron

Caftron—*New Automatic Semiconductor Test Equipment. (In use at C.P. Clare booths 2218-20 as well) *New precision AC-DC Voltage Comparators. Multiconductor Cable Test Sets. Automatic Test System Modules for Ground Support. Production Testing, Incoming Inspec-tion, and Quality Control.

Be sure to see all four floors!

Ortho Filter Corp., Booth 1626 7-11 Paterson St. Paterson 1, New Jersey

George G. Pagonis, ▲ Jerome Potash, Wm. Mc-Gravey, Tom Fogg

Complete line of filters—low, high, band pass up to 450 mc. Toroids, magnetic amplifiers, MIL-T-27A transformers. Precision type at-tenuators. Magnetic components designed to meet customer's specifications.

Oryx Company, Booth 4111 13804 Ventura Blvd. Sherman Oaks, Calif. Bernard L. Cahn, Jack Fields, Jack Simon, John Nolan, L. K. Ingber, C. H. Mitchell John Nolan, L. K. Ingber, C. H. Mitchell The only complete line of precision miniature soldering irons, From 6V to 24V, AC, DC, Wgt. ½ oz. to 34 oz. For production, services, or laboratory applications. "Model 115-10, 10 watts, 355° C, 110VAC iron uses 362'' tip. "Model 115-15, 15 watts, 380° C, 110VAC uses 362'' tip. Approved by Department of Building & Safety, Los Angeles.

> John Oster Mfg., Co. **Avionic Division** 1 Main Street Racine, Wis. Booths 1330-1332

▲ Andrew Barbaccia, Howard Driver, Mort Last, George Lathrop, Jack Pinner, Robert Ramm, Donald Uhen, David Yonis



Hi-Temp Size 8 Synchro

Missile quality components. Newest designs in high and low temperature servos, damped servos, synchros (transmitters, receivers, differentials, control transformers), resolvers, generators, mo-tor tachometers, ac drive motors, de motors, motor-gear trains, servo torque units, electro-mechanical assemblies, and indicators.

Oxley Developments Company Ltd., Booth 1820

See: British Radio Electronics Ltd.

Ozalid Division, General Aniline & Film Corp., Booth 4133

Ansco Rd. Johnson City, N.Y.

Walter Berthold

Ozalid Whiteprint Machines for reproduction of engineering drawings and prints; Machines and Materials for Pre-Printed Wiring Circuits, Dia-grams, Production Patterns, etc. Accessories.

PCA Electronics Incorporated 16799 Schoenborn St. Sepulveda, Calif. Booth 1333

John Kane, Charles Rubin, Max Shaw, Andrew Jones

Engineering and manufacture of pulse Engineering and manufacture of pulse transformers, delay lines and toroids. Stock available on pulse transformers and delay lines. Fulse Transformers: For blocking oscillator, coupling, transis-tor circuit application to various packag-ing configurations. Delay Lines: Dis-tributed constant types, lumped constant types, to commercial & military specifica-tions, Toroids: To commercial, military requirements. requirements.

▲ Indicates IRE member. Indicates new product.

PSP Engineering Co., Booth 2229 See: IMC Magnetics Corn

Pace Electrical Instruments Co., Booth 3009

See: Precision Apparatus Co., Inc.

Pacific Automation Products, Inc., Booth 4527

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Glendale 1, Calif. D. Louro, E. Regan, R. Veloz, E. Heimer, E. Albertson, H. Somer

Custom Electronic Cable, Cable Assemblies and Harnesses, Instrumentation Systems, Control Systems, Electronic Installations, Mobile Mounted Equipment, Special Electronic Devices, Checkout Equipment*, Systems Integration, Communications Systems and Architect and Engineering Services,*

Pacific Division, Daystrom, Incorporated, Booths 1704-1706 9320 Lincoln Blvd.

Los Angeles 45, Calif.

Jack H. Zillman, Bob Wolin, Allan Richards, Joe O'Callaghan, Frank McClure

Precision Wire Wound Potentiometers, Square-trim Potentiometers, Sub-Miniature Potentio-meters, Gyros, Pressure Transducers, Altimeters, Flight Control Systems.

Pacific Semiconductors, Inc., Booths 2742-2744

10451 West Jefferson Blvd.

Culver City, California

S. L. Spiegel, Frank E. O'Brien, Robert T. Reid, ▲ H. N. Sachar, Philip Astra, Thomas Gracie, Cal Tainter, Richard Bossert, Henry DiMond, Hal Nodiff, ▲ John Black

Very High Frequency and Very High Power silicon transistors, High Speed Switching tran-sistors, "Micro-Diodes" and Micro-Transistors, High Q Varicaps (voltage-variable capacitors) and conventional silicon diodes and rectifiers.

Packard-Bell Electronics Corp., Defense & Industrial Group, Booths 3707-3709 12333 West Olympic Blvd.

Los Angeles 64, Calif.

▲ R. B. Leng, \blacktriangle G. Bieging, M. Palevsky, A. R. Adams, D. B. White, A. Randazzo, \blacktriangle J. Campbell, H. Miller, \blacktriangle B. L. Soule, \blacktriangle G. Rus-sell, T. Smith, J. Behr

seti, i. Smith, J. Behr *Multiver M.3: Eleven bit bipolar analog to digital converter—all solid state—approx, 10 KC conversion rate; *DC Differential Amplifier No. 361: 200 KC bandwidth—all solid state in-cluding chopper; Aerotape: Airborne tape re-corder-reproducer; ATC Transponder Test Set; *CNI Test Set: Portable, battery operated; RDM Volumeter; Electronicrometer; *Trans-ducer; High temperature, high line pressure to 800°F.

Paco Electronics Co., Booth 3909 See: Precsion Apparatus Co., Inc.

Page Communications Engineers Inc., A Subsidiary of Northrop Corporation, Booth 3932

Page Communications Bldg., 2001 Wisconsin Ave., N.W. Washington 7, D.C.

A Walter Brehm, ▲ Stuart Hyans, ▲ Donald Palmer, ▲ John Redden, ▲ Arnold Rosenberg, Charles Breeding, ▲ H. H. Schenck, ▲ Ross Bateman, ▲ P. D. Rockwell, ▲ Gail Bogg, ▲ William Collins, J. P. Gaines, Charles A. Parry

Worldwide telecommunications systems: under-seas, over the ground, trans-horizon and into space.

(Continued on page 336A)

A complete listing of all registrants at the IRE National Convention and Radio Engineering Show, showing company affiliation and local hotel, is posted on the back wall of the mezzanine on the west side of the first floor.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Northeast Scientific Corp., Booth 2844 30 Wetherbee St., RFD 1 South Acton, Mass

▲ Clement Moritz, John J. Hogan, John T. Parkhill

Regulated High Voltage Supplies with maximum voltages to 10 kv and output currents to 1 ampere, Constant Current Supplies, and Millimeter Wave Generators,

Northeastern Engineering, Inc., Booth 3108

25 S. Bedford St. Manchester, N.H.

▲ C. N. Chagaris, ▲ C. J. Kannair, Jr., ▲ B. E. Lamere, ▲ J. L. McCluskey, ▲ N. L. Westlake Development and manufacture of Frequency, Period, and Time Interval Measurement Instruments and Systems, Frequency Conners and Converters, Frequency Standards, *Digital Printers, 'Code Converters, Preset Decades and Decade Sealers, Inertial Gyro Test Equipment.

 Northern Radio Co., Inc. 143-7 West 22nd St. New York 11, N.Y. Booth 3510
 S. A. Barone, A. A. J. Odgers, A. F. C. Lambert, A. J. S. Harris, A. P. Flamholtz Transistorized Voice Frequency Telegraph Carrier Equipment AN/ FGC-61

Northrop Corp., Booth 3932 See: Page Communications Engineers, Inc.

Nothelfer Winding Laboratories, Inc. 220 Ewingville Road (P.O. Box 155) Trenton 3, N.J. Booth M-13 John J. Nothelfer, Joan Suleskey, Paul Walshin Gapless core transformers, control reactors—maximum KVA, minimum losses and magnetizing currents. Highest efficiency in power and instrument transformers. Toroidal magnetic circuits at cost of production transformers. VARI-HENRY adjustable inductance—TAK-A PART transformer, reactor, saturable reactor—Current transformer through type single and multi-ratio—Toroidal Cores, grain oriented, silicon-steel—Frequencydoubling transformer.

Oak Manufacturing Co., Booth 2921 1260 North Clybourn Ave. Chicago 10, Ill.

▲ L. H. Flocken, H. Howell, P. Parasoe, R. McTigue, W. Cochrane, P. Wheaton, H. Olson, E. Olenick

Low Power Rotary, Slider, Lever and Pushbutton Actuated Switches, Including New Line of "Stock" Switches, AC/DC Choppers, TV Tuners, Variable Capacitors, Rotary Solenoids, Vibrators, Appliance Timers.

> Offner Electronics Inc. 3900 N. River Road Schiller Park, Ill. Booth 3051

George W. Little, \triangle Dr. Franklin F. Offner, James Janisch, Richard Cozak Type R Dynograph multichannel directwriting oscillograph, all transistor, one microvolt de sensitivity and convertible recorder. Also 12 to 19 channel writer console, de differential data amplifiers and other recorders.

(Continued on page 334A)

▲ Indicates IRE member. * Indicates new product.





from 10 KC to 600 MC

MODEL 91-CA 300 microvolts to 3 volts Price: \$495

MODEL 91-C 1000 microvolts to 3 volts Price: \$395



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Ultra-Reliability

is our watch word at GUIDANCE CONTROLS CORP.



SI7E 0 POT.-CLUTCH MODULE

Pancake style, 2.5 watt clutch coil. Rated Torque: Mag. ener-

gized ... 16 oz. in, Mech. energized . . . 8 oz. in. Pot. wiper integral with clutch shaft, zero backlash, zero axial play. Specs, for GCP-9 same as listed below for Precision Pots. & Magnetic Clutches.

MINIATURE GEARHEAD MOTOR, SIZE 11

Ratios from 5:1 to better than 78,000:1 as required. Operating Load Torque . to 100 oz. in. max. υp Backlash: 30 minutes max, for any ratia. Gearing: Precision Class II Bearings: ABEC Class 5 Shaft End Play: .002" max. meas-ured with 1 lb load. Shaft Radial Play: .0005" max. measured with 4 oz. load.



PRECISION POTENTIOMETERS A complete line of linear and non-linear potentiometers. LINEARITY . Standard .08% Best .03% MAXIMUM **RESOLUTION .015%**

Low torque, law naise, excellent stability. Exceeds MIL 5272 and NAS 710.

MAGNETIC

CLUTCHES Unique electra-magnetic design pravides far extreme tarque ta inertia ratio...aver 1,000,000 rad./sec².

SPECIFICATIONS

- Zera backlash and .0005" max. end play Friction face ABEC-5 ball bearings .0015" TIR input/autput shaft cancentricity with servo maunt pilat
 Exceeds MIL-5272, 4158, and 8189 specs.

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I.R.E. SHOW . . . BOOTH M-2



Most advantageaus for critical size/critical weight applications 14 madels available in sizes 6 and 8

SUB & MICRO MINIATURE

MAG. CLUTCH BRAKES

High tarque ta inertia ratia. Zera backlash and end play.
Exceeds MiL-E-5272,

4158, and 8189 specs.

SEND FOR FREE DATA SHEETS GUIDANCE



Whom and What to See at the Radio **Engineering Show**

(Continued from page 329A)

New Hermes Engraving Machine Corp. 154 West 14th St. New York 11, N.Y. Booth 4035 N. Schimmel, H. Susskind, W. Dannheisser, G. Berlant, M. Kaufman, R. H. Laird



Engravograph

Portable and hench type engraving machines, tracer-guide for unskilled labor. Special equip-ment for drilling printed circuits. Ultrasonic industrial cleaning machines. Nameplate bevel-ler, Cutter grinder, Laminated engraving stocks. ENGRAVOGRAPH, the pantograph engraving machine for nameplates, dials, panels, routing & milling on all plastics & metals—unlimited dimensions.

New York Air Brake Co., Kinney Vacuum Division, Booths 4309-4311 See: Kinney Vacuum Division

Nicad Division, Gould-National Batteries, Inc., Booth 1116 172 Pleasant St.

Easthampton, Mass.

W. R. Albrecht, F. C. Anderson, A. H. Lind-say, L. R. Mannheim, R. T. Perron, R. L. Ringer, T. Ulrich, J. D. Whittemore, Jr., R. L. Wilkinson

NICAD nickel cadmium batteries are reliable, compact, light in weight, have low resistance and provide sustained voltages at high dis-charge rates over a wide temperature range. They recharge rapidly, withstand shock, vibra-tion and indefinite storage. No corrosive fumes -low maintenance.

Non-Linear Systems, Inc. Del Mar Airport, Box 728 Del Mar, Calif. Booths 3041-3042 Andrew F. Kay, Richard C. Wynne, Peter Van Benschoten, Ben Fisher, Rob-ert Landay, Thomas E. Nawalinski

Operate-it-yourself display of digital Voltmeters and voltage comparators for research, inspection, standards labs, auto control, data loggers. Modern packaging permits changing instruments parts in 10 minutes. Introducing new digital at price of laboratory type pointer meters. Precision resistors.

Norbute Corp., Booth 2134 See: Kurman Electric Co.

Norrich Plastics Corp., Booth 4046 206 Babylon Turnpike Roosevelt, L.I., N.Y., Richard E. Thaw, Philip Brody Epoxy Bolbins, Epoxy Coils, Epoxy Potting Capsules, Plastic Screw Machine Products, Precision Metal Servo Machine Products, Molded Plastic (Epoxy) Parts, Precision Metal Screw Machine Products.

North American Electronics. Inc. 71 Linden St. West Lynn, Mass. Booth 2009

Raymond W. Smith, Charles H. Trout, William C. Cacciatore, Paul Norton, Dick Henry, James Wall, Arthur H. Bruno, Avio DiFelice, Paul Flaherty, Peter Whoriskey, Joseph C. Miller Development and production of semicon-ductors to standard and special specifica-tions to meet customer requirements. Standard product lines include miniature diodes, low and medium power rectifiers. 150 MW to 50 watt avalanche diodes. 50 mA to 20 anp. silicon controlled rectifiers.

North Atlantic Industries, Inc. 603 Main Street Westbury, L.I., N.Y., Booth 3833 ▲ Malcolm D. Widenor, ▲ Walter Lipkin, Sid-ney Herman, John A. Gregorio, ▲ Herbert W. Bass, ▲ J. P. Brogan, ▲ E. V. Donegan, ▲ E. Brown Brown



*New Model RB-504 Ratio Box

Single, Multiple & Broadband Frequency Phase Angle Voltmeters, Portable & Militarized, Pre-cision AC Inductive Voltage Dividers & Stand-ards, Complex Voltage Ratiometers, Modular AC-DC Phase Sensitive Converters, Oscilloscope, Ratio Test Sets, Servo Driven Self Balancing Indicators (Military & Commercial Applications) with Repeating Indication of Temperature, Milli-volts, Ratio, Synchro, etc.

North Electric Co., Booth 2125 553 South Market St. Galion, Ohio

koetter

H. H. Brewer, W. F. Keally, J. V. Guercio, P. Van Valkenburgh, W. F. Tidd, H. R. Rivitz, **& W. W.** Crissinger, C. V. Schuster, T. W. Parsons, R. E. Pickett, G. Hinkle, R. Rosen-

System Techniques utilizing North products such as Supervisory Control, Electronic Switch-ing, Automatic Controls, Transistorized Tone and Voice Communications Equipment, Ampli-fiers, Relays, Crossbar, Rotary Stepping Switches, Connectors, Transistorized Power Supplies, Electric Counters.

North Hills Electric Co., Inc. 402 Sagamore Ave. Mineola, L.I., N.Y. Booth 3023

Sydney Cramer, Leo Staschover



Equipment Department: Constant Current Suppiles for testing gyros, diodes, transistors, re-lays, magnetic cores, batteries, fuses, bolometers, accelerometers. Components Department: Coils, toroids, filters, wide band transformers. The new Sanborn Model 320 system — for general purpose DC recording in any part of the plant or on field assignments — combines rugged currentfeedback amplifiers, 2-channel recorder assembly and dependable all-transistor circuitry in less than a cubic foot of space. And the many advantages of Sanborn multi-channel systems are incorporated in the new portable 320 — low impedance, enclosed galvanometers; clear, permanent traces made by heated styli; rectangular coordinate charts. Most components for each channel are mounted on one easily serviced card; others are readily accessible. The control panel permits easy access to the controls for each channel ... provides for observation of 6 inches of the chart ... and it can be set up for use vertically, horizontally or at a 20° angle using the adjustable stand/carrying handle.

Your nearest Sanborn Sales-Engineering Representative can provide you with complete data or write the main office in Waltham. Sales-Engineering Representatives are located in principal cities throughout the U. S., Canada and foreign countries.

DIRECT WRITING SYSTEM



- up to 0.5 millivolt/mm sensitivity
 - inputs floating and guarded for each channel
 - rectangular coordinate charts full 50 mm wide
 - only 12¾" square, 8¾" deep
 - 4 pushbutton chart speeds
- completely transistorized

S P E C I F I C A T I O N S

electrical
Sensitivity Ranges 0.5, 1, 2, 5, 10, 20 mv/mm and v/cm
Input Impedance $1\!\!\!/_2$ megohm on mv, mm ranges and 1 megohm on v/cm ranges
Frequency Response 3 db down at 125 cps, 10 div peak-to-peak
Common Mode Voltage = 500 volts maximum
Common Mode Rejection
Linearity maximum non-linearity 0.2 mm with respect to chart center
Calibration
Limiting approx. $\pm 115\%$ of full scale
Rise Time 4 milliseconds with less than 4% overshoot
physical
Input Connectors
Output Connectors 40 mv/mm sensitivity for connection of external monitoring scope to each channel
Dimensions
Weight

Controls

(for each recording channel)
Range Switch 6 positions and off
Smooth Gain Control
Function Switch 4 positions - Zero, Cal, Use mv/mm
Position Control Use v/cm
Stylus Heat Control

Galvanometer Damping Galvanometer Compensation (screwdriver adjusts.)

Data subject to change without notice.

See this new System at the I.R.E. Show-Booths 3601-03-05



PORTABLE 2-CHANNEL

022

MENI

National Co., Inc., Booths 1401-1407 61 Sherman St.

Malden 48, Mass.

S. L. Rudnick, E. R. MacDonald, R. H. Rogers, ▲ F. Roberts, S. W. Natkin, J. H. Quick, H. C. Guterman, ▲ E. Grant, ▲ S. Fast, ▲ P. Smith, ▲ J. Hannigan, ▲ Hal Moyer, ▲ Earl Sloane, ▲ Stan Turner

Radio Receiver equipment, Electronic compon-ents, Elasticable, Servo Motors, Generators and Systems, ATOMICHRONS, and Military honospheric and Tropospheric Scatter Com-nunications Equipments and Components.

National Research Corp., Booth 4427 See: NRC Equipment Corp.

> National Semiconductor Corp. Sugar Hollow Road Danbury, Conn. Booth 2007

▲ Dr. B. Rothlein, Dr. E. Clarke, J. Gruber, Dr. R. Rau, R. Hopkins, G. Schneider, J. Hegarty, B. Wonnacott, Dr. M. Schneider, D. Harris, R. Hunchak



Silicon Transistors: PNP Alloy =2N1440 series & 2N327A series, highest device dissipation (400 nw at 25°C, 170 nw at 125°C), surpassing all multistry specifications, for small signal and medium power applications; NPN Mesa- 2N702 series & 2N560 for switching and computer applications cations

National Union Electric Corporation, Electronics Division, Booth 1328 Bloomington, Ill.

▲ E. R. Ewald, B. J. Hart, H. B. Graham, D. E. Bartmess, ▲ W. R. Schweikert Special purpose vacuum tubes, high voltage regulator, miniature cathode ray tube, wide band secondary emission amplifiers and muneral glow readout tubes.

Navigation Computer Corp., Booth 3223 1621 Snyder Ave. Philadelphia 45, Pa.

J. Paul Jones, Jr., David M. Biber A Henry Longley, Jr., Henry Apfelbaum Biberman, Transistorized digital systems modules, includ-ing Shift Registers, Binary Connters, Reversi-ble Counters, Decade Counters, Digital Delays, Switches, Gates, Digital to Analog Converters, NOR Logic, etc. Featured will be a new Numerical Read-Out Device.

Nems-Clarke Co., Div. Vitro Corp. of America 919 Jesup-Blair Drive Silver Spring, Md. Booths 3826-3828 ▲ A. S. Clarke, ▲ R. E. Grimm, J. F. Whitehead, K. B. Redding, R. P. May, R. C. Curry, P. Dudney, M. L. Bandler, T. W. Maskell, D. H. Steinweg, C. Hall, W. H. Kimbell Complete Line of RF Telemetry Equip-ments. Receivers, Preamplifiers, Multi-couplers, Spectrum Display Units, Range Extension Units, Jacks, Jack Strips and Broadcast Items.

New Haven Clock & Watch Co., Booth 2240 See: Condenser Products Div.

(Continued on page 332.4)



Possess all of the quality and dependability for which the McCoy line of metal encased crystal units is famous.

Check these advantages:

Excellent Long-Term Stability

Minimum Aging

Choice of Leads -Pins or Flexible Wire

Maximum Resistance to Shock and Vibration 30 vector G's from 20 to 2000 cps — vibration 100 G's — shock

True Hermetic Seal Altitude is no problem

Meets new CR-73/U and CR-74/U Specs

Wide Range of Frequencies Available 5000 KC to 200.000 KC

Extremely Small Size



now available in a wide range of STANDARD SIZES





precision tooled beryllium copper

for: Transistor, Capacitor, Diode, Fuse and **Component applications.**

B T I electronic clips are now stocked for prompt shipment in a complete range of standard sizes and designs. The use of beryllium copper and associated alloys insures positive spring contact pressure with exceedingly high electrical and thermal conductivity.

Write today for the new series of **B T I** bulletins showing standard sizes with specifications.



grounding strips and ring contacts for all electronic requirements. Write for Bulletin #E-106.

BRAUN TOOL & INSTRUMENT COMPANY, INC.



Whom and What to See at the Radio **Engineering Show**

(Continued from page 326A)

Muirhead & Co., Ltd., Booth 3230 Beckenham, Kent, England See: Muirhead Instruments, Inc.

Muirhead Instruments Inc. 441 Lexington Ave. New York 17, N.Y. Booth 3230 ▲ J. V. Foll, J. A. Muirhead, ▲ P. L. Irvine, A. E. W. Hibbitt, A. Cooper, L. W. Fenn, ▲ A. Roy Smith, ▲ F. A. Miles, E. A. Conte, L. F. Purcer



Automatic Analyser Equipment

*Automatic Wave Analyser, 10 c/s to 19k c/s for automatic analysis of complex noise and vibration waveforms. Wide range Decade Oscil-lator 1 c/s to 100k c/s. High Frequency Analyser-Oscillator 200 c/s to 650k c/s. *Spe-cial type Synchros and Motor Generators. Wes-ton Standard and Reference Cells.

Murata Manufacturing Co., Ltd., International Division, Booth 2006 Kaiden, Nagaoka-Cho

Otokuni-Gum, Kyoto, Japan Tatsuya Akebi, 🛦 Osamu Saburi

Ceramic Capacitors, high temperature and close tolerance. Subminiature type Tuners for applica-tions in radio, T.V., and Computers. Latest de-yelopments in Mechanical Filters and Tuning Forks

Mycalex Corp. of America 125 Clifton Boulevard Clifton, N.J.

Booths 2729-2731 Edward de Villeroy, George Lynch, Winfield Darrow, William Ormston, Thomas Weber, John Froemel, Francis Barr, Henry Richardson, A. Monack, Richard Young

MYCALEX® glass-bonded mica preci-sion-molded tube sockets, arc shutes and high temperature electrical components and parts. SUPRAMICA® ceramo-plastics—telemetrv switches, commutator plates. RF coil supports and high tem-perature electrical components and parts. SYNTHAMICA® synthetic mica— pow-ders, paper, crystals in various forms.

▲ Indicates IRE member. * Indicates new product.

A complete listing of all registrants at the IRE National Convention and Radio Engineering Show, showing company affiliation and local hotel, is posted on the back wall of the mezzanine on the west side of the first floor.

NRC Equipment Corp. 160 Charlemont St. Newton Highlands 61, Mass. Booth 4427

Booth 4427 J. H. Moore, R. H. Binkerd, S. G. Bur-nett, H. M. Farrow, G. King, Jr., D. D. Preis, J. J. Flood, B. Thorley "Multiple-source Laboratory Vacuum Coater. "High Speed (1550–1/s) 6" Fractionating Diffusion Pump. "High Performance Mercury Diffusion Pump for lamp exhaust. "Nottingham UHV (to 10-10mm, Hg.) Ionization Gauge Control. "Redhead UHV (to 10-14mm, Hg.) Gauge and Control. "Battery-Op-erated Thermocouple Gauge. 'Extreme Ahitude Radio-Sonde.





Voltage Power Pulse Modulators and High Voltage Power Supplies. Catalog Item Versatile Modulators for Magnetron and Klystron Pulse Operation. Test Modula-tors and Power Supplies for Laboratory and Production Testing.

Narda Ultrasonics Corp., Booth 4532 625 Main St.

Westbury, L.I., N.Y.

A Dr. John C. McGregor, Walter H. Venghaus, Paul W. Steen, Strain Sutton, Norman Pad-den, Richard Fallon, ▲ Robert Markel, ▲ Mar-tin A. Damast, ▲ Morris Kenny, ▲ Eugene Black

Showing a complete line of ultrasonic cleaning equipment from 35 watts to 3 kw, including transducerized tanks as well as immersible transducers. Inquiries on custom installations are invited. Showing for the first time a com-pletely transistorized unitized ultrasonic cleaner.

National Carbon Co., Booths 2401-2403 See: Union Carbide Consumer Products Co.

National Cash Register Co. Electronics Div. 1401 E. El Segundo Blvd. Hawthorne, Calif. Booth 1912

L. J. Hines, W. Irwin, E. J. Haley L. J. Hines, W. Irwin, E. J. Haley A line of computer memory components. Cylindrical thin film electro-deposited magnetic rod; magnetic rod array, High speed magnetic rod memory; Ferrite core array for memory applications meeting cxtreme environmental conditions; micro-encapsulation and photochromism.

328A



A new corporation devoted exclusively to the design, fabrication and installation of antenna systems in the fields of scatter communications, missile tracking, space tracking, radar and surveillance, radio astronomy and special antenna products.



A word from the new president, Chorles Creaser... "Ours is primarily an engineering organization which is employee-owned and employee-run. Our objective can be stated simply: it is to lead in the development and introduction of new and improved techniques and processes. We are equipped to handle the entire

antenna system — reflectors, mounts, feeds, pedestals, waveguide, rotary joints — everything from transmitter and receiver on."

For more information, please send for our folder



... and from the vice-president, Bill VonderWolk "We're off to an exciting start. We've taken a new approach to antenna marketing by building a new, 30-foot parabolic dish for space tracking and communications, which promises to be more accurate than anything

yet built. Very soon, we'll have the finished product, built and operating, to show to industry and government. Instead of offering a design and a promise of performance, we'll prove ours first.

Hingham Industrial Center, Hingham, Mass.



World Radio History

Whom and What to See at the Radio Engineering Show

(Continued from page 324A)

Model Engineering & Mfg., Inc., Booth 2305
See: Tru-Ohm Products Div.
Moeller Instrument Company, Inc., Electronics Division, Booths 2005
132nd St. & 89th Ave. Richmond Hill 18, N.Y.
A Charles Beck, Jr., A John B. Chatterton, A David G. Hollister, Joseph A. Reale
Isolated dc Power Supplies, Power Isolators, Isolated dc Power Supplies, Power Isolators, Isolated Line Voltage Sources for instrumentation and testing. Engineering. research and manufacturing facilities for any instrumentation isolation problems.
Molectronics Corporation, Subsid. Minneapolis Moline Co., Booth 3952

neapolis Moline Co., Booth 3952 1717 North Potrero Ave.

South El Monte, Calif.

Josh Gershuny, Harry Robertson, ▲ Joseph Steinfeld, Charles Stewart

Electronic test equipment and instrumentation; digital voltohmmeters, digital ratiometers, digital counter units, vacuum tube voltmeters, oscilloscopes, relay testers, and custom instruments; portable and rack mounting units.

Show Hours

10 a.m. to 9 p.m. daily Monday through Thursday March 21-24, 1960 Molecu-Wire Corporation Eatontown-Freehold Pike Scobeyville, N.J. Booth 42:12 A Edward E. Edmunds, A Stephen Poch, Arthur S. Lichter, Albert Young Fine and superfine resistance wire .0004" to .010" diameter. Bare, enamelled, oxidized and fabric covered finishes. Also precision grade Grid Lateral wires.

Monroe Calculating Machine Co., Booths 1610-1618, 1709-1717 See: Litton Industries, Inc.

Monsanto Chemical Co., Booths 4026-4027

800 N. Lindbergh Blvd. St. Louis 66, Mo.

Monsanto Organic Chemicals Div.: F. H. Langenfeld, R. A. Steenrod, A. G. Eades, Denis Perry, R. Draper, P. G. Benignus, M. McEwen Coolanol 45 coolant-dielectric for electronic equipment. Aroclor series of dielectrics for capacitors. transformers. Radiation-resistant, other synthetic fluids.

Monsanto Inorganic Chemicals Div.: ▲ R. F. Meehan, ▲ John Weber, R. A. Staniforth, J. J. Burrage

Semiconductor Products Including-"Ultra-pure zone-refined silicon single crystal rods; "polycrystalline silicon rods; intermetallic materials.



F. L. Moseley Co., Booths 3210-3212 P. O. Box 791, 409 North Fair Oaks Pasadena 3, Calif.

▲ Francis L. Moseley, ▲ James H. Burnett, Myron H. Hunt, Robert N. Flanders, Glenn Whiteley

X-Y Recorders, Strip Chart Recorders, Logarithmic Converters, AC to DC Converters, Continuous Line Followers, Magnetic Curve Followers, Punched Tape and Card Converters, Servo Voltmeters, Magnetic Tape Converters.

Donald P. Mossman, Inc., Booth 2337 P. O. Box 265

Brewster, N.Y.

Donald P. Mossman, Jr., Ben Rist, Ed Brunner, Alan MacLachlan, Robert Maher, Wm. Brabant, Herb Kahn

Lever Switches, multiple contact. 3 to 20 Ampere, Push Button Switches, Interlocking, momentary, "Alternate action, Illuminated, "Series 5900 and 7700 now available with detachable contact assemblies to facilitate changes in contacts, simplify stocking. Dust covers to enclose pileups also available.

Motorola Inc. Semiconductor Products Division 5005 East McDowell Road Phoenix, Ariz. Booths 1114-1115

DOOTHS 1114-1115 A Richard H. Rudolph, \triangle F. Joseph Van Poppelen, \triangle James LaRue, \triangle A. B. Dall, \triangle H. I. Ackerman, \triangle Charles F. Scott, William A. Kraus, David P. Hall, Robert R. Thomas, Raymond G. Kimbell, Edward J. Loyd, Arthur Powell, \triangle Glen Madland, G. E. McGonagle, \triangle Larry Kelly, James Lucy, John L. Gray, Jerry Sanders



New 1/4 Watt glass silicon Zener Diodes

Motorola presents important design information, reliability data and working applications on UHF Mesa Transistors; Industrial Alloy and Power Transistors; Silicon Rectifiers (including automotive); and the industry's most complete line of Silicon Zener Diodes, reference elements and glass diodes.



Subminiature ceranic capacitors tailored to your space requirements. Various ceranics for choice of properties. Ribbon, wire and tab lead arrangements. Low-voltage, high-capacitance units for transistor circuitry. NARROW-CAPS for 1/10 inch mounting.

(Continued on page 328A)

▲ Indicates IRE member, * Indicates new product.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

Measure transistor characteristics—

Alpha (h_{fb}) Beta (h_{fe}) Input Resistance (h_{ib})

-with the new BRC Type 275-A Transistor Test Set



See us at the RADIO ENGINEERING SHOW BOOTHS 3101-3102

Exclusive BRC features –

- Unique null-type measuring circuit completely unaffected by signal level fluctuations
- Reads Alpha to three significant figures
- Accurately measures at extremely low emitter currents

– plus

- Permanent Calibration
- Direct Readout of Alpha, Beta, and Input Resistance on large easy-to-read dial
- Built-in adjustable, metered collector and emitter
 power supplies
- Handles up to 5 amperes emitter current
- Measures both NPN and PNP transistors
- Special test circuit guards against transistor burnout

Precision Electronic Instruments since 1934



BOONTON . NEW JERSEY . U.S.A.

World Radio History

The BRC Type 275-A is an exceedingly flexible and efficient instrument for the precise measurement of basic transistor parameters over an extended range of operating conditions. It can also be used to measure the characteristics of diodes and other semi-conductor devices. Direct readout of the following parameters -

- Alpha (hfb) Beta (hfe)
- Input Resistance (hib)

is presented on a large, easy-to-read dial without correction or interpolation. Two built-in, fully regulated, low ripple power supplies furnish completely variable emitter current and collector voltage.

Alpha Measu	urement (ha):	
RANGE: ACCURACY:	(a) 0 to 0.99 (b) 0.9 to 0.999 (a) $\pm 1\%$ (b) $\pm 0.5\%$	
	•when $fa \stackrel{>}{=} 500$ Kc.	
RANGE: 0 to	rement (h _{fe}): 200	
Input Resist RANGE: ACCURACY:	ance Measurement (h _{1b}): (a) 0.3 to 30 ohms (b) 3 to 300 ohms (c) 30 to 3000 ohms (a) ± 3% (b) ± 3% above 30 ohms* * for linear impedances	
FREQUENCY ACCURACY:	t Oscillator: : 1000 cps ± 5%	
Collector Vo	Itage Supply:	
External: External:	0 to 100 V.D.C. 0 to 100 V.D.C.	
Range: Accuracy	0 to 2, 5, 10, 50, 100 volts $\pm 1.5\%$ full scale	
Emitter Cur	rent Supply:	
Internal: External:	0 to 100 ma D.C. 0 to 5 amp. D.C.	
Ranges:	0 to 0.1, 0.2, 0.5, 1, 2, 5, 10, 20, 50, 100 ma. + 15% full scale	



Electronic. Electrical. **Mechanical Components** and Contacts with **NO Film or Residue**





APPLICATIONS

Electronic Components & Assemblies: Diodes, Transistors, Slip-Ring Commutators, Crystals, Vacuum Tube Components, Sub-Miniature Assemblies.

Meter & Instrument Components: Instrument Bearings, Jewel Bearings & Pivots, Gear Trains, Lapped Surfaces.

Electrical Contacts: Relays, Vibrators, Voltage Regulators, Sensitive Switches.

FEATURES

No film, residue, or corrosive effect to damage surface, fire and explosion hazard nil, non-polar, non-ionic, an all around safe operation.

For specific information about your critical cleaning problems, send product information and production requirements.



See us at the IRE show—booth #4106

Whom and What to See at the Radio **Engineering Show**

(Continued from page 323.4)

Mincom Division Minnesota Mining & Mfg. Co. 2049 South Barrington Ave. Los Angeles 25, Calif. Booth 3923 ▲ R. J. Brown, J. E. DeLand, C. S. Tobias New Instrumentation Magnetic Tape Recording/Reproducing Equipment.

Minneapolis-Honeywell, Boston Division, Booth 2208 40 Life Street

Boston, Mass.

George Bailey, Richard S. Burwen, Leonard P. Entin, Donald G. Lynam, Frank Melanson, John W. Rhinesmith

John W. Kninesmith Gyros; M.100, Golden Gnat (cutaway), JRT, "Star of Gyros" panel (Gnat three axis pack-age for F-106), Accelbata series d-c ampli-fiers, T6GA Galvanometer amplifier, 2HLA-9 Differential huput Indicating Amplifier, 2HCT Precision Temperature Controller.

Minneapolis-Honeywell, Davies Division, Booth 2214

10721 Hanna St.

Beltsville, Maryland J. D. Mitchell, G. A. Lacas, J. G. Booth, N. S. Bassett, H. H. Barnes, R. M. Hadley, R. O. Hutchinson, C. J. Clarke

Analog Direct and FM Record and Playback System, Loop Transport, Wave Analyzer, Rec ord and Playback single and multi-channel Head Display.

Minneapolis-Honeywell **Marion Electrical Instrument** Division **Grenier** Field Manchester, New Hampshire Booth 2202 E. S. Maury, H. A. Donahue, Jr., E. E. Doherty, Gordon Steady, Richard Rat-tigan, Herbert Lawrence, Jr. AC Iron Van Panel Meters, Medalist Meters, Ruggedized Meters, Aircraft In-struments, Miniature Indicators and

Mechanisms.

Minneapolis-Honeywell, Micro Switch Division, Booths 2204-2206 Freeport, Illinois

John Tropsa, Gerald Boyle, William Betz, Rob-ert Remley, Sidney Doctor, Robert Schrader, Jack Ellis, Arnold Bahnsen, Alfred Bahnsen, Sylvan Markosian, Walter Landers, Robert Eadie, Eugene Moran, Donald Guide

Laute, Lugene moran, Donald Guide Bifurcated-contact subminiature switch, "One-Shot" pushbutton assemblies, Snap-acting push-button switches, Alternate-action pushbutton switch, Rapid-repeat, light-touch pushbutton. Decimal-to-binary rotary input switch, Sub-miniature door inter-lock switch, Minature tog-gle switch, and Environment-free sub-miniature switch.

First and Second floors—Components

Third floor—Instruments and Complete Equipment

Fourth floor—Materials, Services, Machinery

Minneapolis-Honeywell Rubicon Division Ridge Avenue at 35th St. Philadelphia, Pennsylvania Booth 2210 Alex Schoemann, Lou Gill, Earl Benson

Wheatstone Bridge, Kelvin Bridge, Potentiometers, Galvanometer, Resistors (NBS and Reichsanstalt).

Minneapolis-Honeywell, Semiconductor Division, Booth 2212 1015 S. 6th St.

Minneapolis, Minnesota

R. O. Anderson, S. L. Furber, R. L. Larsen, L. T. Macgill, R. E. Mock, W. R. Rittman, M. C. Walker

Germanium power transistors; triodes and tetindes.

Minneapolis-Moline Company, Booth 3952

See: Molectronics Corp.

Minnesota Mining & Mfg. Co., Magnetic Products Div., Booth 1913 900 Bush Ave.

St. Paul 6, Minn.

Dan Denham, Wm. F. Enright, James L. Kam-iske, Charles A. Alden, J. Bimrose, J. Rogers, P. Van Deventer, John Watson, Robert Patter-son, P. J. Cafferty, J. F. Maye, J. G. Bondus "Scotch" Brand Instrumentation Tape (Mag-"Scotch" Brand Instrumentation Tape (Mag-netic) and Accessories.

Minnesota Mining & Mfg. Co., American Lava Corp. & Mincom Div., Booths 4502 & 3923

See: American Lava Corp. & Mincom Division

Mitchell-Rand Mfg. Corp., Insulation Div., Booth 4123

51 Murray St.

New York 7, N.Y.

John R. Mitchell, Jr., William A. Ingraham, John J. Finn, Joseph Konkolosky, Harry Bych, Don Pennett, William Lange, Frank Whelan Epoxy Resins, Impregnating Waxes, Potting Compounds, General Line of Electrical Insulat-Compounds, G ing Materials.

> Mitronics Inc. 1290 Central Ave. Hillside, N.J. Booth 1815

Harvey Pensack Smith, Bob Eaves, Tony Monari, Allan Davis, Lloyd Lamb, Jerry Lynch, George Resetter, Billy Schulz, Ed Domber, Jerry Mullen, Norman Mullen, Herb Thode, Carl Engle, Paul Hollenbeck, Hart Vancroft, Lloyd Moore



*New Sub-Miniature Metallized Housing

High alumina ceramic metallized, high tempera-ture applications, molybdenum, manganese, plated with nickel, silver, or sintered gold. *High temperature ceramic precision, thicknesses low as .005" specializing semiconductor and vacuum tube fields. High temperature multi-pin headers. Ceramic to metal brazed rectifier housing assem-blies. Solder Seal terminals. Component housings.

(Continued on page 326A)

▲ Indicates IRE member.

* Indicates new product.

324A

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Microwave Electronic Tube Co., Inc., Booth 2008

76 Lafayette St.

Salem, Mass. ▲ Richard J. Broderick, Harold F. McEnness, ▲ Harold Heins, Philip A. Bagnell, ▲ Louis W. Roberts, Milton R. Hamilton

Development and manufacture of Microwave Duplexing Tubes and Devices—Gas Switching, TR and ATR tubes, Magnetrons, Klystrons, Waveguide Components, Spark Gaps, Pressuriz-ing Windows, Complete Duplexing Assemblies.

Mid-Century Instrumatic Corp., Booths 3837-3839

See: Computer Systems, Inc.

Mid-Eastern Electronics, Inc., Booth 3009

32 Commerce St. Springfield, N.J.

Gunther A. Bielefeld, Lawrence C. Oakley, ▲ William W. Hartz, ▲ Roy L. Anthony, Rich-ard E. Weber, Jud Williams, Michael S. Cold-well, ▲ Alvin Rymsha

ST Series of 19 New Transistorized Power Supplies Featuring Extreme Reliability & Ease of Servicing; Model 801 Ultra High Resistance Bridge; "New Series of Plug-In Constant Volt-age & Constant Current Transistorized Power Supplies Programmable From A Distant Point; "Model 710H Megatrometer.

Military Systems Design, Booth 4307 See: Instruments Publishing Co., Inc.

James Millen Mfg. Co., Inc. 150 Exchange St. Malden 48, Mass. Booth 2517 ▲ James Millen, ▲ E. E. Williams, ▲ R. Wade Caywood, Owen I. Haszard, Philip A. Eyrick



Encapsulated Inductors, Grid Dip Meters, An-tenna Bridges, Bahnis, Electronic Equipments and Components; Magnetic Metal Shields; De-lay Lines, Delay Line Kits, Complete Jine sub-miniature components including mechanical de-vices and coils; Instrumentation Oscilloscopes.

Millivac Instruments Div. of Cohu Electronics, Inc. 2315 Second Ave., Carman P.O. Box 997 Schenectady 3, N.Y. Booth 3607 ▲ Dr. W. K. Volkers, Mrs. D. J. Volkers, Imek Metzger, Donald P. Morey



World's first ultra-low-noise amplifier which combines high input impedance (10 meg) with "hushed transistor" operation less than 0.5 microvolts).—Low-drift chopers.—Linear scale ohumeters and bridges.—Sensitive VTVM-s.

(Continued on page 324A)

PROCEEDINGS OF THE IRE March, 1960

the fastest



fastest rise time of all sampling oscilloscopes -0.4 mµs rise time $-0.05 \text{ m}\mu\text{s/cm}$ fastest sweep speeds - to 300 mc w. model 603 trigger unit fastest rep rates and

highest sensitivity

easiest to use

most versatile

- up to 2.5 mv/cm (30:1 SNR at full scale deflection)
- no critical adjustments
 - complete line of options and accessories

WORK

АТ

- There are more Lumatron Millimicrosecond Sampling Oscillo-scopes in use than any other type . . . at Raytheon, NRL, Bureau of Standards, Texas Instruments, Transitron, Hughes, General Transistor and many others.

UMATRON

LUMATRON - FIRST IN MILLIMICROSECOND INSTRUMENTS See the complete line of instruments for all fast rise time applications AT IRE - BOOTH 3059 or write for complete information and new Millimicrosecond Engineering Data Chart P-3.



Lumatron Electronics, Inc., 68 Urban Ave., Westbury, N.Y.







 .0005% SENSITIVITY
 6-DIAL DECADE
 POWER SUPPLY AND DETECTOR INCLUDED
 UP TO ±.01% FOR
 FULL SCALE DEFLECTION

Operates as Go-No-Go Limit Bridge from 100 to 11,111,100 ohms with *full scale* tolerance selection of 0.01, 0.02, 0.05, 0.1, 0.25, 0.5, 1.0, 5.0, and 10%. Wheatstone Bridge from 1 ohm to 111 megohms. *Write for Bulletin P-100*.



Whom and What to See at the Radio Engineering Show

(Continued from page 320A)



R. E. Tucker, E. F. Shine, John Santillo Advanced design Micro-Matic Precision Wafering Machines with the new *Roton Table Drive, Micromech Diamond Wheels, Semiconductor crystal holding and Orientation Fixtures, Optical Orientation systems, and the revolutionary new Micromech *"Performeter" (Diamond wheel performance indicator).

Microtran Co., Inc. 145 E. Mineola Ave. Valley Stream, L.I., N.Y. Booth 2315 ▲ Albert J. Eisenberg, ▲ Harold Edelstein, ▲ Richard Chaber, Walter Benscher



Miniature transistor transformers consisting of ultra-miniature, subminiature low level chopper, input, DC-DC converter, silicon rectifier, power supply, transistor output, driver, powers, chokes, military and industrial calibre. Custom designed transformers and full line of miniatures stocked at franchised distributors.

One registration entitles you to permanent entry to the show for all four days. Be sure to keep your identification badge or pocket card and bring it with you when you return. Registration is not transferable. Microwave Associates, Inc. Northwest Industrial Park Burlington, Mass. Booths 2301-2303 A Dana W. Atchley, Jr., George S. Kariotis, Richard Dibona, A Eric Stromsted, Robert J. Allen



X-Band Tunable Magnetron, MA-219

TR, ATR, PRE-TR and Receiver Protector Tubes, Ferrite Isolators, Circulators, and Complete Ferrite Duplexers, Microwave Mixer and Video Diodes; *New High Voltage Varactors and *New "Pill" Varactors; Computer Diodes; Microwave Limiters and Switches (Coax). Waveguide Components (Including *New Frequency Multipliers) X-K-Band.

Microwave Development Labs., Inc. 92 Broad St. Wellesley 57, Mass. Booth 2407 A Dr. Henry J. Riblet, A Nathaniel Tucker, A James R. Corcoran, A Kenneth D. Jeffries, A Edward Salzburg, A John D. Hall



Broad Band Balanced Mixer

Research, Development, and Manufacture of Microwave Components and Assemblies, including Adapters, Attenuators, Bends, Fixed Tuned Detent Cavities, Couplers, Crystal Holders, Diplexers, Duplexers, Flanges, Filters, Generators, Hybrids, Balanced Mixers, Monopulses, Rotary Joints, T's, Ferrite Circulators, Tuners, and Phase Shifters.

▲ Indicates IRE member. * Indicates new product.



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TYPE

CY uuf for uuf, the smallest, most stable axial lead capacitor you can buy. Probably ½ smaller than you're used to. After load life tests at 125° with 150% of rated voltage. average change in capacitance is less than 0.4% for 1,000 hrs., less than 0.6% for 10,000 hrs. They exceed all requirements of MIL-C-11272A.

Medium-power transmitting style

CYYF Fusion sealed. Similar to CY, but with glass encapsulation fusion sealed to capacitor and leads to make seal tight against moisture and corrosives. Insures reliable performance under extreme environmental conditions. Guaranteed four times better than MIL specs for moisture resistance.

Wafers with or without leads. Smallest high stability capacitor available. Up to 10,000 uuf in .061 sq. in. of PCB area. Electrodes sealed to dielectric sheets in such a way that seal cannot be broken without destroying capacitor. Meets the performance requirements of MIL-C-11272A.

High temperature dielectric and radiationtolerant metal electrodes with tab leads. Dielectric strength is twice rated voltage applied from one to five seconds. Insulation resistance in ohm x farads is 100 at 175° C., 25 at 250°, C., 1 at 300° C., and .05 at 350° C.

		SIZE	CAPACITANCE (uuf)	DCVW	тс	MAX. CASE SIZE
		CY10	1 to 150 151 to 240	500 300	140±25ppm/°C. from55°C.to +125°C.at	11/ ₃₂ x 11/ ₆₄ x 5/ ₆₄
	Service.	CY15	151 to 510 511 to 1,200	500 300	100 10 11110	¹⁵ / ₃₂ x ¹ %4 x %4
<u></u>	ISI K	CY20	511 to 3,300 3,301 to 5,100	500 300		47%4 X 27%4 X %4
		CY30	3,301 to 6,200 6,201 to 10,000	500 300		4%4 x 3%4 x %4
		CY60	Up to 56,000	Ratings to 4000 peak	140±25ppm/°C.	1 x 1½ x 5%
	T. Same	CY70	Up to 150,000	volts Ratings to 6000 peak volts	$+125^{\circ}$ C. at 100 kc or 1 mc	1½ x 1¾ x ¾
		CYF10 CYF15	1 to 150 151 to 240 151 to 510 511 to 1,200	500 300 500 300	140±25ppm/°C. from - 55°C.to +125°C.at 100 kc or 1 mc	11/32 X 11/64 X 5/64 15/32 X 11/64 X 7/64
W WI		W, WL5 W, WL4 W, WL3 W, WL2 W, WL1	1 to 560 561 to 1,000 1,001 to 2,700 2,701 to 4,300 4,301 to 10,000	300 300 300 300 300	140±25ppm/°C. from — 55°C.to +125°C.at 100 kc or 1 mc	.281 x .218 x .090 .281 x .312 x .090 .531 x .312 x .090 .531 x .453 x .090 .531 x .812 x .090
		НТ1 НТ2 НТ3	1 to 1,000 1,001 to 3,000 3,001 to 10,000	300 300 300	0-250°C. 115±25 0-300°C. 140±35 0-350°C. 160±45	1/2 x 3/8 x 3/6 1/2 x 5/8 x 3/6 1/2 x 5/8 x 3/6 1/2 x 1 x 3/6

Why you have to smash these Corning capacitors to affect their reliability

Stack alternating layers of glass ribbon and aluminum foil, fuse the stacks under heat and pressure, and you have a solid, practically indestructible capacitor.

The properties of the capacitor are *entirely* those of the closely controlled dielectric. They cannot be altered in processing. They stay the same under heat, moisture, and all other environmental conditions.

There's no problem with delivery. We mass produce them all.

If you need capacitors high in reliability, small in size, and light in weight, you should know more about this Corning design. The coupon will bring you complete technical data. Address: Corning Glass Works, 542 High St., Bradford, Pa. For orders of 1000 or less, contact your distributor serviced by Erie Resistor Division.

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Name
Company
Address
City Zone State

Whom and What to See at the Radio **Engineering Show**

(Continued from page 318A)

Measurements A McGraw-Edison Division P.O. Box 180 Boonton, N.J. Booths 3501-3503 ▲ Harry W. Houck, Walter B. Manson, Jr., C. Edwin Williams, Albert B. Eld-ridge, ▲ Norman W. Gaw, Jr. ridge, ▲ Norman W. Gaw, Jr. Standard Signal Generators, Transistor Testers, Frequency Meters, Pulse Gen-erators, Square Wave Generators, Tele-vision Signal Generators, UHF Radio Noise & Field Strength Meters, Megohm Meters, Peak Voltmeters, R.F. Attenua-tors, Crystal Calibrators, Intermodula-tion Meters, Megacycle Meters, Induct-ance Bridges, Capacitance Bridges, Vac-uum Tube Voltmeters, Special Test In-struments. struments.

Mechatrol Div. of Servomechanisms, Inc., Booth 2812

See: Servomechanisms, Inc., Mechatrol Div.



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Mezzanine at back of first floor, South Room at center of south wall, second floor. 3000 court at southeast corner of third floor. 4000 court at southeast corner of fourth floor, 4500 court in northwest corner of fourth floor.

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Menlo Park Engineering 711 Hamilton Ave. Menlo Park, Calif. Booth 3843 ▲ Harold W. Harrison, ▲ John B. Pettegrew, Frank Abel



Traveling Wave Tube Ampliher

Madium power broadband microwave amplifiers. Low noise broadband receivers and microwave oscillators, These units are available to cover the range 0.5 to 18.0 KMC in six frequency bands. Both solenoid and permanent magnet focused units are available for the low and medium power broadband amplifiers.

Mepco, Inc. 35-37 Abbett Ave. Morristown, N.J. Booths 2802-2804 L. Kirby, Jr., E. L. Beaudry, Jr., J. E. Mervin, ▲ R. Gebhardt, C. Miller, ▲ C. Engert, W. Randolph, J. Feller, R. Halpin, N. Delker



Manufacturer of Accurate and High Stability Wirewound and Film Resistors. Wirewound, Carbon Film, Metal Film, Meter Multipliers, Power Resistors, and Custom Resistor Net-

Merck and Co., Inc. Electronic Chemicals Div. Rahway, New Jersey Booth 4513

C. A. Graf, Dr. G. Krsek, M. Judge, C. Carroll, William Reed, Dave Hall, Dr. T. Benedict, F. Bourassa

All forms of Merck Ultra Pure Silicon including float zone refined doped single crystals in various diameters and "new lengths; vapor deposited Single Crystal-line Silicon layers of different types and resistivities; III--V compounds; and ther-moelectric material.

▲ Indicates IRE member. * Indicates new product.



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Metals & Controls Division, Texas Instruments Incorporated, Booths 1409-1421

See: Texas Instruments Incorporated, Metals & Controls Division

Methode Manufacturing Corp., Booth M-3

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Chicago 31, 111. W. J. McGinley, W. T. Jensen, W. E. Stub-bings, A. D. Weiss, J. N. Barrett Connectors—Miniature AN Per MIL-C-26500, "RELI-ACON" Printed Circuit Card Recepta-cles Per MIL-C-21097, Rack and Panel Con-nectors with Closed Entry Contacts. Custom Printed Circuitry for Military and Commercial Applications. "PLYO-DUCT" Flexible Wiring, Tube Sockets and Tube Shields for Standard and Printed Wiring Applications.

Metronix, Inc., Booths 3916-3918 See: Assembly Products, Inc.

Micamold Electronics Mfg. Corp., Div. General Instrument Corp., Booths 1218-1224

65 Gouverneur St.

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▲ E. Geoghegan, A. DiGiacomo, F. Senft, M. Lissner, C. Feinsot, J. Imperial, W. Pelliccia, B. Kohl, ▲ I. Clarke, A. S. Gartner, S. Solomon

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Micro Gee Products, Inc., Booth 3846 6319 W. Slauson Ave., P.O. Box 1005 Culver City, Calif.

▲ B. W. McFadden, ▲ Walter E. Peterson, David S. York

David S. fork *Model GOA Oscillating Rate Table for fre-quency response testing of Gyros and Accelero-meters, 0.1 to 150 cps, 100 pound load capacity. Model 10C Flight Simulation Table for physical simulation with analog computers.

Micro Switch Div., Booths 2204-2206 See: Minneapolis-Honeywell

Microdot, Inc. 220 Pasadena Ave South Pasadena, Calif. Booths 2101-2103

▲ Guy M. Martin, Jr., Robert S. Dickerman, Forrest Besocke, Philip Steward, Edith Lurie, Barry Tunick, Maureen Rainen



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Micro-miniature coaxial connectors and cables, assemblies; ultra-miniature connectors; low noise "mini-noise" cable; miniature coaxial SPDT switch*; custom designed transformers*.

(Continued on page 322A)

NEW FOR FAST, ACCURATE, CONVENIENT FREQUENCY MEASUREMENTS

Microline DIRECT READING

FREQUENCY METERS

Sperry has now developed a new family of Direct Reading Frequency Meters. These new instruments, covering the S, C, X, U, and K bands, provide the ultimate in convenient frequency measurements while still maintaining the accuracy necessary for most applications.

These frequency meters use a revolutionary steel tape drive for fast, accurate reading; misinterpretation is eliminated since only a small segment of the readout tape appears under the window at any given setting.



Model No.	Frequency Range (Kmc)	Absolute Accuracy	Reaction Type Q	Scale Length (in.)	Smallest Scale Division (Mc)	Price
1251	2.60-3.95	.08%	12,000	82	1	\$450
12C1	3.95-5.85	.08°°	10,000	82	2	\$250
12X1	8.2-12.4	.08°₀	8000	80	5	\$150
1201	12.4-18.0	.1 %	7000	81	10	\$210
12K1	18.0-26.5	.1 °0	6000	113	5	\$220

AT THE NEW YORK IRE SHOW

See these new direct reading instruments, along with Sperry's complete line of high precision micrometer frequency meters. We'll be looking for you at Booths 2432 to 2438.



SPERRY MICROWAVE ELECTRONICS COMPANY, CLEARWATER. FLORIDA . DIVISION OF SPERRY RAND CORPORATION

Whom and What to See at the **Radio Engineering Show**

(Continued from page 316A)

McCoy Electronics Co. Mt. Holly Springs, Pa. Booth 2311

Jacoby, Luther W. McCoy, David B. Jacoby, George K. Bistline, Jr., Edward P. Boise, John A. Sward, Donald E. Kutz Precision built quartz crystal units, her-metically scaled in metal or *all glass*^{*} holders; bandpass and single sideband crystal filters, crystal discrimutators, crystal ovens, and switching assemblies.

McDowell Electronics Inc., Booth 4128 105 Forrest St. Metuchen, N.J.

R. B. McDowell, D. Day, S. J. McDowell, Vincent Wagner, Jim Borbely

RF Induction Heating Equipment, Variable Speed Turnable*, RF Current & Voltage Meas-urement Equipment*, Glass to Glass Scaling urement Ea Equipment 5.

QUARTZ CLOCK

TYPE B-288

Description The B-288 quartz clock consists of 6 standard units mounted on slides in a case of painted

non-corrosive metal. It is very solidly built and

is designed to give direct access to all parts of the quartz clock. For maintenance purposes,

any unit can be replaced in less than one

An emergency power supply can be added;

it is also mounted in a case of painted non-

corrosive metal and is entirely independent

The 6 units are connected as shown in the diagram given below. They contain the fol-

McGraw-Edison Co., Booths 2737-2741 & 3501-3503 See: Bussmann Mfg., Div., T. A. Edison Indus-tries, Measurements Div.

McGraw-Hill Book Co., Inc. 330 W. 42nd St. New York 36, N.Y.

Booth 4405

Zoe Gregory

Scientific and Technical Bool: Publishers in the fields of electronics, nucleonics, and control engineering.

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Production engineering is rapidly becoming one of the most important facets of the radio-electronic industry. Fourth floor exhibits are devoted to this subject. On the fourth floor, you can find new machinery and tooling techniques, and discover ways to make your own product better, cheaper and faster. Don't pass up this opportunity to acquire new methods and knowledge on that all-important aspect of electronics-

PRODUCTION—FOURTH FLOOR

McGraw-Hill Publishing Inc. 330 West 42nd St. New York 36, N.Y. Booths 4314-4316

Booths 4314-4316 ▲ James Girdwood, ▲ W. W. MacDonald, A John M. Carroll, ▲ Bruce A. Winner, R. S. Quint, F. M. Stewart, Hugh J. Quinn, ▲ Wm. Hodgkinson, ▲ D. Miller, ▲ H. Shaw, ▲ G. Werner, D. Furth, ▲ W. Gardner, ▲ B. Boyle, D. Watson, ▲ S. Weber, ▲ R. Charest, F. Leary, ▲ M. Tomaino, H. Janis, S. Carter, Wm. O'Brien, J. Mason, ▲ W. Bushor, T. Emma, S. Vogel, L. Solomon, ▲ M. Perugini, G. Flynn, ▲ M. Wolff, N. Lindgren, H. McKean, ▲ H. Hood, ▲ H. Harris, ▲ T. Maguire



Electronics—published weekly for more than 52,000 research, design, production and man-agement people to provide editorial coverage in engineering, design, production, use, business statistics, markets, news of component parts, ac-cessories, equipment, circuits, and complete sys-tems. Electronics Buyers' Guide—published an-nually.

McLean Engineering Labs. P.O. Box 228 Princeton, N.J. Booth 1624 Wallace W. McLean, ▲ W. Benjamin Ecken-hoff, ▲ James G. Robinson, A. Donald Hay



Compact McLean Packaged Blower Cooling devices for electronics and missiles,

McMillan Laboratory, Inc. Brownville Ave. Ipswich, Mass. Booth 3237





All Steel Prefabricated Microwave Free Space Room

Airport ramp testing hoods for aircraft radar-Microwave absorbing accessories for antenna in-stallation-Radiation attenuating limings for shielded or free space rooms.—*Power line radio interference filters-VHF-UHF Microwave anechoic test chambers Microwave testing inter-ferometers and thickness gages.

(Continued on page 320A)

B.+ (B+ (B)

Power supply and synchronous clock Phase-shifter, 50 c/s Oscillator, 100 kc/s Frequency divider, from 100 kc/s to 1 kc/s Frequency divider, from 1 kc/s to 50 c/s Frequency-converter, 200 c/s to 60 c/s 2. 3. 4.

Uses

The complete equipment of the B-288 quartz clock opens up a very wide field of applications. such

1. A high precision chronometer, with rate variations of less than 1/100 sec. per day. It is useful for measuring and testing very short periods of time as well as for timekeeping over long periods. 2. A permanent timekeeper, for, fitted with the emergency power supply, the clock is unaffected by accidental variations or temporary breakdowns in the mains supply. It can therefore be used to govern a master clock distributing the exact time.

3. A frequency standard which is easily transportable yet extremely reliable. The standard fre-quencies of 100 kc/s, 10 kc/s, 1 kc/s, 200 c/s, 60 c/s and 50 c/s can be distributed throughout a building without intermediate amplification.

Thanks to the very high phase-stability, the accuracy of the 50 and 60 c/s outputs is as good as that of the 100 kc/s, oscillator. The 60 and 50 c/s signals and the impulses given each second can control precision devices used in industry, scientific research, navigation and the army. 4. An industrial timekeeper which can be used in watch- and clockmaking to synchronise appa-ratus for testing the instantaneous rate of time-pieces or to control seconds-timers.

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minute.

of the clock itself.

lowing elements.



Selecting a duplexer for high-power applications involves consideration of peak power, average power, transmit loss, receive loss, expected life, and versatility of operation.

All Microwave Associates high power gas duplexers utilize special window structures for optimum switching efficiency without sacrifice in low-level loss characteristics. These windows insure reliable, long-life performance. Both our gas and ferrite duplexers may be operated over very broad bandwidths at the common microwave frequencies.

Exceptionally complete ultra-high-power design and test equipment is utilized by our Research and Production Departments. Each duplexer is fully tested at maximum rated power before shipment.

We have extensive experience in designing and manufacturing high-power duplexer devices and are interested in working on newest ultra-high-power applications. We are now developing ultra-high-power duplexers for more efficient switching at UHF, L, C, and S bands.

Our Applications Engineers would like to discuss the future of high power duplexing with you.

Frequency Band	Duplexer Type	Peak Power	Average Power	Transmit Loss (max.)	*Receive Loss (max.)	Bandwidth
	Gas	5 Mw	300 Kw	0.1 db	0.4 db	Tunable
UHF	Gas	25 Mw	75 Kw	0.1 db	0.4 db	1
L	Gas	25 Mw	50 Kw	0.1 db	0.5 db	
	Gas	6 Mw	30 Kw	0.1 db	0.7 db	
S	Ferrite	3 Mw	5 Kw	0.5 db	0.9 db	10% Nominal
c	Gas	5 Mw	5 Kw	0.1 db	0.7 db	
	Ferrite	5 Mw	7.5 Kw	0.3 db	0.8 db	
x	Gas	500 Kw	500 W	0.2 db	1.0 db	
	Ferrite	1 Mw	1 Kw	0.3 db	0.9 db	
	Gas	150 Kw	150 W	0.2 db	1.0 db	
Ku	Ferrite	150 Kw	150 W	0.3 db	0.9 db	
Ka	Ferrite	75 Kw	75 W	0.3 db	1.1 db	4% Nominal

LEADS THE

ULTRA-HIGH-POWER

INDUSTRY IN

DUPLEXING

with both GAS DISCHARGE

and FERRITE DUPLEXERS

All Microwave Associates duplexers incorporate low-loss, long-life, receiver protectors which guarantee crystal protection over wide temperature ranges and under extreme environmental conditions.

*The duplexer receiver loss includes the loss due to receiver protector TR tubes.

At each frequency band of the microwave spectrum, Microwave Associates has devices for efficient switching of high power.



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MICROWAVE ASSOCIATES, INC.

BURLINGTON, MASSACHUSETTS Western Union FAX-TWX: Burlington, Mass., 942 • BRowning 2-3000

Whom and What to See at the Radio **Engineering Show**

(Continued from page 314A)

Magnetic Instruments Co., Inc. Subsid. Pyrometer Co. of America, Inc. 637 Commerce St. Thornwood, N.Y. Booth 3017 ▲ Robert Levine, Kenneth Lord, Leo Lindell, Richard Levine, Jack Metzger, Leonard Bonn



*Low Level DC Amplifier, Model 759-6

Rate of Change Computer, "New Capacitance Level Gauge, "A New and Inexpensive Low Level D.C. Amplifier, "New Low Voltage, Low Impedance Power Supply,

Magnetic Metals Co., Booth 1432 Hayes Ave. at 21st St. Camden 1, N.J.

▲ H. F. Porter, ▲ D. O. Schwennesen, W. J. Miller, W. G. Pettit, H. V. Paynter, G. L. Mor-row, D. O. Stanton, ▲ W. T. Mitchell, H. H. Hackett, W. Y. Hallman

Electromagnetic cores and shields; Centricores made from Super Square mu "79" will be fea-tured. These tape wound cores are identically duplicate in all essential magnetic dimensions from unit to unit in any lot and from lot to lot without deviation.

Magnetic Shield Div., Booth 4312 See: Perfection Mica Co.



Magnetics, Inc., Control Div., Booth 2437

See: Control, Div. Magnetics, Inc.

Magnetics Specialties, Inc., Booth M-13 See: Nothelfer Winding Laboratories, Inc.

Magtrol, Inc., Booths 1230-1232 See: Travco Associates

D. E. Makepeace Division, Booths 2110-2118 See: Engelhard Industries, Inc.

* Indicates new product. ▲ Indicates IRE member.



P. R. Mallory & Co., Inc.

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Booth 3213 ▲ J. M. Shapiro, H. Feldman, S. Jacobson, M. Seroy



High Power Pulse Modulator (top) Crystal Frequency Synthesizer (bottom)

*Crystal frequency synthesizers; *Highly stable oscillators and reference frequency generators, kilocycles to kilomegacyles; *Ultraprecise crystal ovens; *High power pulse modulators; *High voltage power supplies; *Microwave communica-tions systems; *High level UHF transmitters; *Custom pulse transformers.

Marconi Instruments 111 Cedar Lane Englewood, N.J. Booths 3301-3305 R. J. Bailey, \blacktriangle W. A. Buck, \blacktriangle S. E. Ellam, \blacktriangle J. L. Garthwaite, B. E. Morris, \blacktriangle W. Oliver



20-300 Mc/S Oscillator

Precision Capacity Bridge, *VHF Impedance Bridge, *FM Signal Generator for Mobile Ra-dio, Tunable Communications Receiver with crystal stability, *SSB Transmission Test Set, *Signal Generators AM/FM, Precision Devia-tion Meters, Noise Loading Test Set, Double-Pulse Generator, Oscilloscopes.

Marion Electrical Instrument Div., Booth 2202

See: Minneapolis-Honeywell

Markem Machine Co., Booths 4210-4212 150 Congress St. Keene, N. H.

John G. Powers, Bernard E. Toomey, A. J. Marshala, Joseph H. Lyon, John E. Kelen, Harold B. Lampman, Gilmore G. Fretz, Roland Benson, George E. Fraser, Jr., William B. Morgan

*New miniature JAN diode printer—*small mo-torized offset printer—*U-1026 semi-automatic transistor side wall printer—122A automatic printer for top and side wall with latest features; 69AC color bander up to 6 colors simultaneously on diodes, resistors and capacitors.

Martin Co. Baltimore 3, Md.

Booths 3919-3921

Peter Roche

Electronic Systems for Air Defense, ASW, Reconnaissance, Guidance, Con-trol, Communications, Support Equipment.

Massa Labs., Div. Cohu Electronics, Inc., Booth 3609-3611 5 Fottler Rd.

Hingham, Mass.

ing System*.

▲ Ernest Massa, Kevin Corbett, ▲ Frank Massa, Jr., Jack Hubbard, Tom Pickett Portable Recorders, Multichannel Recording Systems, Amplifiers, Sound Pressure Micro-phones, Accelerometers, Hydrophones, Vibration Exciters, Ultrasonic Transducers, Depth Find-ing Transducers, BNA-250 Recording System^{*}, BSA-650 Recording System^{*}, BSA-850 Record-ion Sourcematic

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Los Angeles 59, Calif.

Al Franklin, Herman Jones, Chuck Sloane, Gene Burroughs, Howard Siegel Roto-Tellite, Word Indicator Lights, Roto-Tel-lite Switch Lights, Time Delay Relays, Phase Sequence Relays and Flashers.

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Potentioneters, precision, wirewound, $\frac{1}{2}$ " to 3", linear and non-linear, high temperature, her-metically sealed. Custom designed to special function, or choose from established standard line. Precision rotary switches and various elec-tro-mechanical devices designed to order.



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Development and manufacture of Air Height Surveillance Radar equipment; microwave antenna components; micro-wave phase shifters; receivers and 1-F amplifiers; electronic scanning radar sys-tems; fuzes; safety and arming devices; analogue computer components and sys-tems; mapping and navigational radar.

W. L. Maxson Corp., Unimax Switch Div., Booths 1204-1206 See: Unimax Switch Div.

(Continued on page 318A)

FIRST AID ROOM First mezzanine. Take elevator 20 from north side of any floor.

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TELE-FLEX...standard moulded sections, the ideal waveguide for use where vibration mounts are not practical.

TELE-TWIST . . . permits quick twisting for immediate field use on "E" and "H" plane bends.

TELE-FORM . . . the finest pre-formed waveguide for use where extremely tight radii must be held.

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NEW, COMPACT RADAR BEACON, designed by TELERAD offers =2 MC/Sec. receiver stability with a Receiver/Transmitter frequency range of 2750 - 2950 MC/Sec. Combined weight of 4.7 lbs.

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Frequency range: 2750 - 2950 MC/Sec. Frequency stability: $\pm 2 MC/Sec.$ Pulse Power: 100 watts peak (min.) Pulse repetition rate: 2000 P.P. Sec. Pulse Width: 0.65 ± 0.05 microsecond Delay: 1.5 microsecond Range Jitter: 0.1 microsecond Size: 2" x 3%" x 7-9/16" Weight: 3.45 lbs. Power Supplies Available on Special Order



NATIONAL REPRESENTATIVES:

Arizona: Premmco-of-Arizona, P. O. Box 663, Scottsdale California: Wallace & Walloce, 1206 Maple Avenue, Los Angeles / G. B. Ellis Sales Co., 2316 El Camino Road, Mountain View / Illinois: Lee Falkenburg, Airborne Sales, 1665 North Milwaukee Ave., Chicago / New Jersey: Telerud Manufacturing Corporation, Route 69, Flemington Pennsylvania, Delaware, Maryland, New Jersey and Northern Virginia: John A. O'Niell Sales Associotes, P. O. Box 311, Collingdale, Pa. / Texas: Southern Industrial Electronics, 429 Exchonge Bldge., Dallas Canada: Instronics, Ltd., P. O. Box 51, Stiltsville, Ontario



RECEIVER Frequency range: 2750 · 2950 MC/Sec. Bandwidth: 6-12 MC/Sec. @ 3 DB Points 35 MC/Sec. maximum @ 40 DB Down Triggering sensitivity: -41 DBM Interrogation: Single or double pulse Frequency stability: ±2 MC/Sec. Size: 11/4 " x 27/6" x 53/4"

SEE US AT IRE SHOW BOOTH 1427

Reliable microwave components and equipment for the electronic, aircraft and missile industries

Weight: 1.25 lbs.



1440 BROADWAY / NEW YORK 18, N.Y. BRyant 9-0892



WHAT'S IN IT FOR YOU?

You may now select from the industry's most versatile and complete line of precision snap-action switches, indicator lights, push-button switches, toggle switches, Switchlites, and environment-free limit switches. You can now make broader product groupings for greater quantity discounts. With this new single source, you will now deal with just one sales engineer for all your switch needs.

Three plant locations-Folcroft, Pa., Chicago, Ill., and El Segundo, Calif.-will provide regional engineering and manufacturing facilities to speed delivery and service.

Local sales offices with factory-trained personnel have been set up to provide on-thespot application engineering in all major markets. An expanded nation-wide distributor organization will assure vou of immediate delivery from local sources.

ELECTROSNAP HETHERINGTON



Whom and What to See at the Radio Engineering Show

(Continued from page 312A)

Litton Industries, Inc. **336 North Foothill Road** Beverly Hills, Calif.

Booths 1610-1618, 1709-1717 Booths 1610-1618, 1709-1717 Grafton P. Tanquary, A Robert C. Kolts, David S. Rathje, Stuart G. Whit-tlesey, Robert Dolbear, David Mutchler, Beverly Kumpfer, John C. Jacobs, James W. Weidenman, Karl Clough, A L. W. Howard, A Ernest Clover, A Edmund Shimbel, Jack T. Gentry, Richard J. Kuri, George W. Steck, Henry P. Bech-told, Jr., Leo Call, Teck A. Wilson, Jack Thorne, Merritt McKnight, Carl Tendler, George Constantine, Anthony Easton, John Shonnard, Vincent Gale, James Martin, Lawrence Truitt, Kenneth McConnell, Herhert Israel, Peter Marzan

Peter Marzan Direct writing picture tube demonstra-tion, computer components, shaft encoder, microwave power tubes, inertial refer-ence elements, communications equip-ment, facsimile transmitting and receiv-jus systems, airborne digital systems, potentiometers, transformers, printed circuits, ferrite isolators, related micro-wave components, terminals & hardware, gas noise sources, display tubes, demon-stration of pulse rate modulation trans-mission. mission.

Lord Manufacturing Co., Booth 2131 1635 West 12th St. Erie 6, Pa.

M. D. Wood, J. M. Weaver, V. Ellis, G. H. Billman, J. J. Goodill, R. P. Thorn, J. P. Cooney

Unit isolators and complete mounting systems (standard and custom designed) for shock, vibration and noise control of sensitive equipment.



▲ Indicates IRE member. Indicates new product.

MM Enclosures, Inc., Booth 4011 111 Bloomingdale Rd. Hicksville, L.I., N.Y.

Michael C. Presnick, Louis J. Wepy, George Boziwick, Gilbert Bassin, Phillip Luce, Carl Berntsen, Robert Ebert, Kenneth L. Reidy, Norman Stachalek, Ernie Williams, Chester A. Milewski, Richard Sager

Electronic Instrument, Transit and Combination Cases Constructed To Conform To Military Spe-cifications MIL-7-945, MIL-7-21200, MIL-C-4150, MIL-7-4734 and MIL-STD-108, Metal Electronic Enclosures and Housings, Custom Fabricated Precision Assemblies and Weldments

Machlett Laboratories, Inc., Subsid. Raytheon Company, Booths 2604-2614 1063 Hope St.

Springdale, Conn.

W. Brunhart, H. D. Doolittle, D. S. Frankel, C. Kirka, R. E. Nelson, M. Rome, G. J. Taylor, C. V. Weden, A. F. Wegener, S. Yanagisawa Electron Tubes; UHF Planar Triodes; Shielded Grid Triodes, Tetrodes & Rectifiers; TV Cam-era Tubes; Scan Conversion Tubes; High Power Vapor Cooled Triodes.

MacLeod & Hanopol, Inc., Booth 3711 10 Roland St.

Charlestown 29, Mass.

A. D. MacLeod, L. Hanopol Capacitance Bridges, and Megohumeters for Vacuum Tube Measurements.

Magnecraft Electric Co., Booth 2525 3352 West Grand Ave. Chicago 51, Ill.

H. D. Steinback, M. S. Steinback, W. J. Gor-man, J. E. Deimel

Relays; Telephone Type Relays, Latching Re-lays, General Purpose AC & DC Relays, Mili-tary Miniature Relays, Sensitive Relays, Her-metically Scaled Relays, Crystal Case Micro-Miniature Relays, New Clear Plastic Enclosed Plug-in Relays, New Low Cost General Purpose Relays.

Magnetic Amplifiers, Inc. 632 Tinton Ave. New York 55, N.Y. Booths 1 127A-1427B Edward Frieling, Milton Cohen, B. Weinstein, Lou Yellin



1300VA 3ø Static Inverter Supply

Universal Power Control Units, Static Sequencers and Programmers, Static Sequen-cers and Programmers, Static Inverters, Tran-sistor-Magnetic Servo Amplifiers, Voltage/Cur-rent Regulators, Magne-Speed Drives, Phase Analyzer Transistor Curve Tracer Instruments.

(Continued on page 316A)

The Radio Engineering Show lasts four days There are four floors in the Coliseum. Why not spend one day on each floor to make sure you see all of more than 800 new ideas?

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



Solves problems with Special Purpose Digital Systems

Special purpose digital systems economically provide solutions to many complex problems in comparison, guidance, decision, evaluation, information storage and retrieval, sorting, computation, and similar needs. For speed and efficiency, a digital system cannot be equalled. Two examples are:

PROBLEM: Wanted, a machine to make decisions involved in library-type operatians (sorting, sequencing, selection) on single and multi-page documents photographically recorded on Eastman Kodak Minicard Film Records. These Film Records measure 16 mm. x 32 mm. Each one can hold up to 2730 bits of binary information, or up to 12 legal size pages with 294 bits of binary information. Binary information is photoelectrically scanned at a rate of 20 cards per second and the film records are sequenced where document pages extend to two or more cards.

SOLUTION: A 3C special purpose digital system meets all requirements. The system incorporates T-PAC {1 megacycle} dynamic digital modules and magnetostrictive delay line serial memories. The manner and use of the 'question' material to select, sort and sequence, is determined by plugboard wiring and control panel switches. Even complex logical relationships between 'question' words and Minicard binary information (43 parallel bits per word in length) are specified on the plugboard. All operations are performed with the time-and-equipment-efficiency techniques characteristic of serial dynamic logic. In practice this system has proven highly successful.

PROBLEM: Data-processing installation required to increase computation speed permitting on-line operation. A major portion of computation time is lost extracting roots in pressure-velocity conversions and correcting for nonlinearity in other transducers.

SOLUTION: 3C Modular Transistorized Random Access Magnetic Core Memory was installed to serve as a "linearizer" or table look-up. After calibration, data is stored in the memory (from magnetic or perforated tape, or other permanent source). A digital transducer signal, fed to the memory, acts as an address signal and the memory output produces the linearized signal. This allows linearization of any desired accuracy in a few microseconds, for direct use in a data reduction computer. Signals from one or several transducers can be linearized at rates well over 100,000 per second with a single 3C Memory.

SEE BOOTHS NO. 3308-3310 IRE SHOW

3C has designed and built a large number of special purpose digital systems during the last eight years. This experience is yours for the asking. Call us for consultation on your system requirements — proposals submitted on request.



COMPUTER CONTROL COMPANY, INC.

EASTERN DIVISION: 983 CONCO'RD ST. • FRAMINGHAM • MASSACHUSETTS BOSTON: CEdor 5-6220 • FRAM: TRinity 5-6185 • TWX FRAM 17 WESTERN DIVISION: 2251 BARRY AVE. • LOS ANGELES 64 • CALIFORNIA GRanite 8-0481 • BRadshaw 2-9135 • TWX W LA 6634

March, 1960



Whom and What to See at the Radlo Engineering Show

(Continued from page 311A)

Lieco, Inc. Syosset Industrial Park 130 Eileen Way Syosset, L.I., N.Y. Booth 2435 A. Zeitz, M. Zanichkowsky, S. Goldstein, M. Goldstein, W. McHugh, S. Glassman, A. Cirolia, F. LeBlanc, T. Hendel



± 360° Sector Scan Rotary Joint

In addition to the broadband-high power RG96/U (above), a complete line of rigid and flexible waveguide assemblies, components and test equipment with UTH waveguide assemblies available including balanced phase Aluma Flex-Guide.

Lindberg Engineering Co. High Frequency Division 2321 W. Hubbard St. Chicago 12, Ill. Booth M-19

Paul Bjork Model LA-VSE-24ZR, zone refining scanner for float zone refining silicon and other semiconductor materials and various types of furnaces applicable to the production of semiconductor devices.

Line Electric Co., Inc., Booth 1107 271 South 6th St. Newark 3, N.J.

S. Tobol, F. C. Corbutt, G. G. Galion, C. E. Pellechio

Relays: General Purpose 5, 10, 15 ampere open and plug in; Latching 5 & 10 ampere, open and plug-in type; Thermal Time Delay; Telephone Type; Spot-lite Indicating Relays; Screw Terminal Motor Starting Relays; Footswitches; Polystyrene Housings; Buzzers, AC & DC, Coils.

Ling-Altec Electronics, Inc., Booths 3802-3804 & 3311

See: Ling Electronics, Inc., & Electron Corp.

Ling Electronics, Inc., Division Ling-Altec Electronics, Inc., Booths 3802-3804

 1515 S. Manchester Ave. Anaheim, Calif.
 Cameron G. Pierce, Charles Theodore, Lew Gillingham, Stan Walters, Robert Lewis
 ESD-ASD 20 Equalizer-Analyzer, Shakers, several newest models*, Rand-o-Matic console equipment and Sine-o-Matic console equipment.

Link Belt Corp., Booth M-10 See: Syntron Company

Link Division General Precision, Inc. Hillcrest Road Binghamton, N.Y.	
Booth 1507 C. Miale, ▲ G. Herzfeld, R. Seals, J. Newell, J. Ritchie, R. Thompson	111111
Digital Plug-In Modules, Analog Build- ing Blocks, A/D-D/A Conversion Mod- ules, Servo Systems Components, Servo Systems, Universal Analog Function Generator, Hi-Speed Digital Function Generator, Data Display and Presenta- tion Systems.	

Liquid Carbonic Div., Booth 3060 See: General Dynamics Corp.

> Littelfuse, Inc. 1865 Miner Street Des Plaines, Ill. Booth 2923

J. D. Hughes, W. A. Clements, H. A. Cornelius, W. J. Henke



3AG Indicating Fuse Post

A New series of Indicating 3AG Fuse Posts provide instant blown fuse indication with high degree of illumination, Available across voltage range of 2½ to 250 volts. Minimum current 20 amps—and new sub-miniature Microfuses from 1/500 amp, through 5 amp, at 125 volts, Fuse measures only .205 dia, X.270 long, Precision engineered with high reliability.

Little Falls Alloys, Inc. 189 Caldwell Avenue Paterson 1, N.J. Booth 4015

Harold M. Malm, James Sacco, O. Veltri, Peter Von der Horst



Beryllium Copper, Bronzes, Brasses, Titanium, Nickel Clad Copper, Nickel Clad Titanium, Solder coated and silverplated wire, Round, Flat, Square and Rectangular Shapes.

(Continued on page 314A)

The letter "M" preceding a booth number indicates that the exhibitor will be found on the mezzanine at the back of the first floor.

Whom and What to See at the Radio Engineering Show

(Continued from page 310.4)

Lepel High Frequency Labs., Inc. 54-18-37th Ave. Woodside 77, L.I., N.Y.

Booth 4236

H. Peterson, H. H. Watjen, C. L. Jennings, A. Vescuso, H. Stiefeling, E. N. Curcio, F. G. Holzhausen, J. Dietz, A. Bellini, P. Capolongo, Harry Hoffmann, G. K. Einhellinger



Model HCP-Floating Zone Fixture

High frequency induction heating equipment for semiconductor work, crystal pulling, floating zone. Model HCP floating zone unit, improved model, the first of its kind to be exhibited. Bench model for heating pretiuned wires. Water recirculator.

> Levinthal Electronic Products, Inc. Subsidiary of Radiation, Inc. Stanford Industrial Park 3180 Hanover St. Palo Alto, Calif. Booth 3003

▲ Eli Goldfarb, ▲ Howard Jessup, ▲ Joseph Swanson, ▲ Robert V. Johnson, ▲ Elliott Levinthal, ▲ Albert J. Morris



60 KW Klystron Transmitter, 225-400 mc

Equipment: Transmitters, modulators, power supplies, medium-high-power pulse transformers and pulse current transformers for radar, communication, tube development, Nuclear: Scintillation crystals, transducers.

Librascope Division, General Precision, Inc., Booths 1501-1503 808 Western Ave.

Glendale, California

R. E. Hastings, \triangle C. K. Krill, \triangle H. R. Davidson, \triangle H. Hemmendinger, J. K. Walker, M. C. Hirsch, G. J. Howard, P. D. Ship

Shaft Position To Digital Encoders including Non-Contact Magnetic'; Contact Type with Size 8* and Size 18* Synchro-Mounting; Oil Filled'; Relay Closure*; 10 Bit Per Turn V-Scan'; 10 Bit Per Turn Self-Decoding, X-Y Plotter.

Licon Switch & Control Div., Illinois Tool Works, Booth 1506 6606 W. Dakin St.

Chicago 34, Ill.

J. B. O'Connor, J. O. Roeser, H. F. Benjamin, P. A. McCullough, N. H. McDowell

P. A. McCullough, N. H. McDowell Precision Snap Action switches and related electro-mechanical devices including a complete line of basic, diecast, enclosed, impulse, subminiature and hermetically sealed units. Sealed and unsealed pushbuttons, toggles, levers, and plunger actuators are also available. Pascol Div.—Rotary D.C. solenoids, solenoid-switch assemblies. (Continued on page 312.4)



JULIE RESEARCH LABORATORIES, INC. 556 West 168th Street, New York 32, N. Y. LOrraine 8-8700 Ultra-Precise Resistors, Networks, and Instruments *potent applied for

stability.



Whom and What to See at the Radio Engineering Show

(Continued from page 308A)

Larson Instrument Co., Booth 3121 24 Orchard St. Tarrytown, N.Y.

L. H. Larson, R. Asen, T. Hendel, M. Lichtenstein, G. Adam, A. Shore, C. Sargeant, F. Bowden, B. Chase, B. McCarthy

Plug-in contact meter indicating controls featuring high speed, accuracy, repeatability. Inkless 30 channel recorders featuring 3600 times magnification in 5 milliseconds, binary, digital, sequence, on-off recording.

Lavoie Laboratories, Inc., Booths 3911-3913

Matawan-Freehold Rd. Morganville, N.J.

A. P. Buckley, Robert James, ▲ Thomas Laugesen, Larry Lippert, Louis Tischler

esen, Larry Lippert, Louis Tischler Improved line precision instruments, designed & produced to provide highest standards of performance & reliability in spectrum analyzers, oscilloscopes, frequency standards, frequency meters, counters, specialized oscilloscope development & production for commercial & military electronic sales. *303 Automatic Impedance Checker. *LA-80 Frequency Counter.

Leach & Garner Company Industrial Division Leach & Garner Bldg. Attleboro, Mass. Booth 4052

Gerald F. Tucci, Fred Dole, Sam Greenbaum, Arthur O. Marcello, Jr., Peter Microulis, B. Hocker



Laminated and solid precious metal contact material. Alloys of gold, silver, platimum and palladium, Tubing, wire sheet; Overlay, Inlay, Toplay, Thrulay, and Edgelay, Silver and gold solders, Non-ferrous and precious metal foil strips, Waveguide tubing, Roll formed shapes.

Leesona Corporation, Booths 4323-4325 P.O. Box 1605 Providence 1, R.I.

W. T. Crocker, C. Zaikowski, I. Marsh, W. Rainford, W. Quinn, J. Halliday

Two #107 automatic coil winders equipped with space-wind, paper-miss detector, auto-slow down and other attachments being run simultaneously by one operator. One #108 semi-automatic coil winder featuring quick set up, hand-feed, paper insulating sheets.

(Continued on page 311A)

▲ Indicates IRE member. * Indicates new product.

See all the exhibits!

Don't miss these important locations—

Mezzanine at back of first floor. South Room at center of south wall, second floor. 3000 court at southeast corner of third floor, 4000 court at southeast corner of fourth floor, 4500 court in northwest corner of fourth floor.



912-914 Westfield Ave., Elizabeth, N. J.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE



FULL coverage in Pulse Instrumentation through MODULAR CONSTRUCTION

Modular plug-in construction adds unparalleled versatility and serviceability to proven EP circuit quality, allows extension of standard instruments to special requirements, and provides the key to rapid, economical fabrication of simple or complex pulse and digital instrumentation systems.

Electro-Pulse currently manufactures 137 standard pulse and digital circuit modules (both tube and transistor types). Over 90 catalog instruments are offered to save you time and money in the generation of fast-rise pulses, pulse pairs, pulse trains, gates, time delays, digital words, programmed current pulses, PPM and PCM codes, etc. Our current comprehensive catalog is yours for the asking.

Various combinations of only eleven basic pulse circuit modules,* when plugged into wired rack frames, make up the four standard pulse generators shown above—

- $3450C .015 \ \mu s$ rise single pulses, 50v into 50 ohms to 2MC, variable durations, delay and waveform.
- 3450C/X-Adds pulse pair and pulse train capabilities to 3450C.
- 3450C/Y—Fast rise, power flip-flop (45v into 470 ohms, Pos. and Neg. outputs), duration to 1 sec., rep rate to 1.7MC.
 - 4120B-Economical fast-rise pulses to 500KC, 35v into 100 ohms.

Write for complete data: Bulletins 3450 and 4120

Representatives in Major Cities Electro-Pulse, Inc.

11861 TEALE ST., CULVER CITY, CALIF. * Phone: UPton 0-9193 or EXmont 8-6764

*Basic modules in photo above:

Time Base, Delay and Width Control, Pulse Forming, Flipflop, Trigger Amplifier, 2 Output Amplifiers, 2 Power Regulators, Rectifier-Filter, and Gating Control, with variations. Also available: Counters, And/Or Gates, Crystal Oscillators, Precision Time Delays, Blocking Oscillators, Mixers, Inverters, Attenuators, Input Amplifiers.

Note, in above photo of 3450CX, the ease with which a single module may be extended on plug-in adapter for service.

> Pulse and Digital Circuit Engineers: Rapidly expanding Systems activity and New Product development at Electro-Pulse have created several attractive openings for qualified engineers. Please send resume to T. C. Ridgway, Personnel Manager.

World Radio History

Whom and What to See at the Radio **Engineering Show**

(Continued from page 306A)

Korfund Company, Inc., Federal Division, Booth 1324

48-15 Thirty-second Pl.

Long Island City 1, N.Y.

Dr. B. K. Erdoss, Donald H. Vance, J. I. Ham-mond, & N. Kfoury, F. Purcell, A. White, Lee Perlow, J. Feigen, F. Kirschner, R. Upton Perlow, J. Feigen, F. Kirschner, R. Upton *Vibration and Shock Isolation Systems for Gyros, Inertial Platforms, Computers, and Other Electronic Equipments; *Broad Temperature Elastomer Mount; *All-Attitude Spring Mounts; *Rotationally Constrained Mountings for Gyros; *"I,O-Q" Structural Coating For Resonant Struc-ture Problems; Slock and Vibration Proof Ship-bing Containers

Krengel Manufacturing Co., Inc., Booth 4227

227 Fulton St.

ping Containers.

New York 7, N.Y.

A. L. Gershon, George A. Feldman, John A. Collins, Jr., Howard D. Gershon, Gilbert M. Bloch, Lloyd B. Thompson, Harold Murray, Dominick Ponzo

Dominick Ponzo Marking Devices; Steel Stamps; Printing Ma-chines, Hand or Antomatic; Engraved DEEP-KUT Stamps; Corporation & Notary Seals; Stencils; Daters; Electric Time Stamps; Checks & Tags; Nameplates; Inks; Pads; Numberers; Inspection Stamps; Rubber Stamps; Self Inkers; Automatic Coders.

Krohn-Hite Corp., Booths 3708-3710 580 Massachusetts Ave.

Cambridge 39, Mass. ▲ George Hite, ▲ Wallace L. Bixby, ▲ Richard Haddad, ▲ John McLaughlin, ▲ Norman Rie-menschneider, ▲ Vernon Hopper, ▲ Bob Mey-ers, ▲ Alan Stern

Ultra-hieh regulation tube and "transistorized power supplies; low distortion, wide-range RC cseillators; ultra-low frequency variable elec-tronic filters; ultra-low distortion power ampli-tier and wide-build direct coupled 10 and 50-wart power amplifiers, with a frequency range from de to one megacycle.





Pressure Contact Terminal Block

Terminal blocks, terminal strips, aircraft & electronic switches, Silicon rectifier holders, power outlets, printed circuit terminal blocks, lighting harnesses, "KLIPTITE" connectors on blocks, solder lugs, marker strips, turret type terminals on blocks. Navy type terminal boards, jumpers, spreaders, pressure-contact terminal blocks blocks.

▲ Indicates IRE member. " Indicates new product.

FIRST AID ROOM

First mezzanine. Take elevator 20 from north side of any floor.

Kupfrian Mfg. Corp. 395 State St. Binghamton, N.Y. Booth 4318 W. J. Kupfrian, C. J. Chase, F. E. Kent, W. Shaw, E. Braddock, T. Kennedy, K. M. Grout, K. L. Bitting, A. Hall



Transistorized Inverter

Stock and design-to-specification transistorized power supplies, cable harnesses, flexible shafts for renote control, and specialized hardware, in-cluding miniature Oklham couplings, universal ioints

Kurman Electric Co. Subs. of Crescent Petroleum Corp. 191 Newel St. Brooklyn 22, N.Y. Booth 2134 Raphael Spiegelman, ▲ Julian Goodstoin, Wal-lace Green, Joseph Lauria, Eric Kriegler, John Scotti, I. Cohen, R. Corenthal



Microminiature Relay

A complete and competitively priced line of re-lays—"off the shelf"—including hermetically scaled, sensitive, telephone, subminiature, micro-miniature, polar, power, antenna changeover, motor starting, RF keying, plate circuit, photo electric, open & plug in types, and dust protected.

LIECO, Inc., Booth 2435 See: Lieco, Inc.

LEL, Inc. **380** Oak Street Copiague, N.Y. Booth 2102

 \blacktriangle James Allsopp, \blacktriangle Charles Baker, \blacktriangle Charles Bissegger, \blacktriangle Walter Hollis, \blacktriangle William Maggio, \blacktriangle Robert Mautner, \blacktriangle David McPherson, \blacktriangle Robert Murphy, \blacktriangle Charles A. Nucbling, \blacktriangle Jack Vigiano



*Solid state microwave mixer-preamplifier, *Transistorized command receiver, *Octave RF amplifier, *Telemetering preamplifier, *Micro-wave receiver, *Transistorized preamplifier. amplifier, *Tel wave receiver,



Lambda Electronics Corp. 11-11 131 Street College Point 56, N.Y. Booths 2318-2320

▲ Lester Dubin, ▲ Simeon Weston, Merrill Simon, ▲ Benjamin Shmurak, ▲ Sol Greenburg, William Kellerman

Featuring a new line of high-current, low-voltage, solid-state power supplies. Also, a wide range of transistor-regulated and tube-regulated power supplies for production and laboratory service.

Land-Air, Inc., Booth 1600 See: Stepper Motors Corp.

> Landis & Gyr, Inc. 45 West 45th St. New York 36, N.Y. Booth 3012

Wiesendanger, H. A. Jenny, W. H. ▲ M. Wiesendanger, Goudy, W. Steinmann



SODECO Print Counter

Electro-magnetic impulse counters; electronic high-speed counters; predetermining counters & high-speed electronic predetermining counters is monodecade counters; bi-directional & bi-direc-tional predetermining counters; printing count-ers & printing counters with date/time printer; add-subtract counters; hour, minute, second counters; pulse generators; instrument trans-formers formers

Langevin Div., Booths 1204-1206 See: W. L. Maxson Corp.

Lansdale Tube Company, Booths 1302-1308

See: Philco Corp.

LaPointe Industries Inc., Booth 2113 155 West Main St. Rockville, Conn.

Dan Malone, Fred Wisk, Al Esten, John Sul-livan, Tom Collella, Warren Anderson, Al Fargo, J. H. Stillbach

Printed Circuitry & Packaged Circuitry, Vari-able Capacitors, Electronic Assembly, Beryllium Components for Inertial Navigation, Small Alu-minum Tubing.

(Continued on page 310A)



The Ideal Approach to SSB... Eimac Ceramic Tetrodes from 325 to 11,000 watts

Generating a clean SSB signal is one thing... amplifying it to the desired power level with stability and low distortion is another. A modern Class AB_1 final amplifier designed around an Eimac ceramic-metal tetrode is the ideal answer to the problem. The Eimac ceramic linear amplifier tubes shown above — the 4CX250B, the 4CX300A, the 4CX1000A and the 4CX5000A — offer the high power gain, low distortion and high stability that is needed for Class AB_1 operation. Each has performance-proved reserve ability to handle the high peak powers encountered in SSB operation. Efficient integral-finned anode cooler and Eimac Air System Sockets keep blower requirements at a minimum and allow compact equipment design. And, all four incorporate the many advantages of Eimac ceramic-metal design, which assures compact, rugged, high performance tubes.

The high performance and reliability of Eimac ceramic tetrodes make them the logical starting point in the design of compact, efficient single sideband equipment.

Write our Application Engineering Department for a copy of the technical bulletin "Single Sideband."

EITEL-MCCULLOUGH, INC. San Bruno, California



Eimac

CLASS AB₁ SSB OPERATION

	4CX250B	4CX300A	4CX1000A	4CX5000A		
Plate Voltage	2000 v	2500 v	3000 v	7500 v		
Driving Power	0 w	0 w	0 w	0 w		
Peak Envelope Power	325 w	400 w	1680 w	11,000 w		
World Radio History						



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THE BEST OF THE JETS... PLUS UNITED'S EXTRA CARE

Whom and What to See at the Radio Engineering Show

(Continued from page 304A)

Kings Electronics Co., Inc. 40 Marbledale Rd. Tuckahoe 7, N.Y. Booths 2821-2823 G. Alterman, ▲S. Jackson, ▲G. Nuremberg, ▲ M. Weissman, A. Ferrari, ▲ D. Davis



Individually Calibrated Connector

Exhibit will be a line of Foantlex and TNC connectors with extraordinary VSWR readings individually calibrated for Specific or Broad Band use.

Kingsley Machine Company 850 Cahuenga Blvd. Hollywood 38, Calif. Booth 4235 George E. Kingsley, A. C. Sheffield, Norman Friedman



Hand Operated Wire Marking Machine

Wire marking machines for placing wiring diagram circuit references on the outer covering of insulated wires and cables, Equipment for production of wire identification markers. Electric ovens for continuous sintering of markings on Teflon TFE insulated wires and tubing.

Kinney Vacuum Division The New York Air Brake Company 3529 Washington St. Boston 30, Mass. Booths 4309-4311

W. B. Mills, J. P. G. Davis, W. G. Small, S. MacKiernan, J. A. Brower, A. W. Livesey, G. Brewer, J. Leonetti, A. M. Messina, R. Denton, W. Crowley, J. Donahue



Ultra High Vacuum Production Units, Floating Zone Refiner, Packaged High Vacuum Pumping Systeth, Complete Evaporator Equipment, Speciel Electronic Depositors, High Vacuum Pumps, High Vacuum Components, ULTEK Ultevac Pumps. KinTel Division of Colu Electronics, Inc. 5725 Kearny Villa Rd. San Diego 12, Calif. Booths 3613-3617 Henry Pannell, & Joe Szewzuk, & Earl Cunningham, & Chek Titcomb, Al Braun, Jack Ja-Quay, & Bill Ivans, Bill Royce



Model 601A AC Voltage Standard

DC digital voltmeters, AC converters, militarized AC/DC digital voltmeters, multi-channel scanners, AC and AC/DC digital voltmeter preamplifiers, digital ratiometers, differential DC amplifiers, wideband DC amplifiers, AC voltage standards, DC voltage standards, electronic galvanometers, closed-circuit television systems.

Kintronic Division, Booth M-23 See: Chicago Aerial Industries, Inc.



Knight Electronics Corporation, Booths 3915-3917

210 S. Des Plaines

Chicago 6, Ill. ▲ M. Bond, L. M. Dezettel, O. Fried, J. Korshak, ▲ G. Mills, L. S. Preskill

Professional quality electronic kits, Hi-Fi and Stereo components, test instruments, amateur gear. All specifications unconditionally guaranteed. Amplifiers, tuners, scopes, communications receivers, intercom, VTVM, portable radios, tube testers. Exclusive features for faster, casier building.



(Continued on page 308A)

Be sure to see all four floors!

306A



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Here is the precision scientists need for today's new measurements of time, speed, direction and distance. Here is the stability to speed space exploration, to broaden the application of molecular electronics, and to aid such advanced studies as that involving Einstein's theory of special relativity. Here is a dramatic break-through in solid state physics design.

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Write us for complete technical data.



Whom and What to See at the Radio Engineering Show

(Continued from page 303A) Kemet Company, Div. Union Carbide

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Corporation, Booth 2405

11901 Madison Ave.

Cleveland, Ohio

2825

14 Prince Pl.

Newburyport, Mass.

Kearfott Division General Precision, Inc. Little Falls, N.J. Booths 1509-1511

Mulliam L. Quigley, George Dulay, Vincent O'Donnell, Jack Robertson, John McDonough, Henry Bloom, Frank Abate, Donald Chiafulio, Robert Wray, William E. Watson, James Rigley, Frank Crowley

Inertial Systems, Gyro Systems & Gyros, Servo System Components including Synchros, Servomotors, Tachometers; Test Equipment; Microwave Components; Ferrite Components; Analog Digital Converters, Mechanical Computers; Radar Test Sets; Marine Electronic Equipment; Electrohydraulic components.

Keithley Instruments, Inc. 12415 Euclid Ave. Cleveland 6, Ohio Booth 3907

▲ J. F. Keithley, ▲ J. L. Gibson, ▲ P. Saint-Amour, ▲ J. Praglin, ▲ A. D. Oliverio, ▲ H. Sohn, ▲ R. Wood



Model 151 Null Detector

*Null Detector, Research Micro-Microanumeter; *Differential Input Electrometer, *Static Meter, (Regulated High-Voltage Supply, Microvolt-Ammeter, Milliohumeter; Multi-porpose Electrometers, Megmegolummeters, Linear & Logarithmic Micronicroanumeters, Log and Period Amplifier, DC & AC Amplifiers,

Kellogg Switchboard & Supply Co., Booths 2510-2520, 2615-2625 See: International Telephone & Telegraph Corp.

Kelsey-Hayes Co., Booth 4002 See: Utica Drop Forge & Tool.



Kepco, Inc. 131-38 Sanford Ave. Flushing 55, L.L. N.Y. Booths 2636-2638

Harvey J. Krasner, Max Kupferberg, George Hill, ▲ Kenneth Kupferberg, Jesse Kupferberg, Jack Kupferberg



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Keystone Products Co. 904-6 23rd St. Union City, N.J. Booth 1911

E. M. Sokolowski, ▲ N. R. Grossner, W. H. Winchell, A. Ollick, E. Holewski



Filter-networks, toroids, magnetic amplifiers, transformers, electronic assemblies to customer requirements. Precision, high-reliability engineering and construction, Facilities for R and D of static magnetic devices. All components available encapsulated, molded or hermetic sealed to MIL specs.

Walter Kidde & Company, Inc., Booth 1925

Belleville 9, N.J.

Truman Young, R. Blake, N. Diepeveen, E. Demers, R. Gilbert, T. Jacobson, R. Langfelder, C. Klein, H. Matarazzo, W. Masnik, A. Jennings, C. Ware, E. Olsta

Jennings, C. Ware, E. Olsta Thermistors, Relays, Converters, Power Supplies, Static relays & contactors for control & power applications, the "Freqconstat" ac to ac power frequency changer, dc to ac and dc to dc power conversion units, high temperature thermistors, rathomes, solid state light flushers, pircruit fire detectors, & related solid state equipment.

(Continued on page 306A)

D. S. Kennedy & Co. 155 King St. Cohasset, Mass.

Booths 2532 & 2637

▲ R. J. Grenzeback, J. C. Cotney, F. R. Hart, D. B. Kennedy, J. J. Flynn, ▲ A. R. Hawes Sumcastings for precision aptennas. Ap.

Spincastings for precision antennas. Antenna products, Waveguide components, Transmission towers for utility companies,

▲ Indicates IRE member, * Indicates new product.

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Julie Research Laboratories, Inc. 556 West 168th St. New York 32, N.Y. Booth 3238

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Kanthal Corp., Booth 4425 Amelia Pl. Stamford, Conn.

Erik Hagglund, ▲ G. W. Eisenbeis, R. C. Bossa, E. J. Chappa, P. E. Beach, J. Chester Bussa, E. J. Chappa, F. E. Beach, J. Chester Thermo-Kanthal. Thermocouple alloys, Fine gauge precision alloys for resistors and poten-tiometers. Kanthal DR and Nikrothal I, with high resistivity and low temperature coefficients. Kanthal Super, molyhdenum disilicide heating clements 1700°C, element temperatures.

> Kay Electric Co. II Maple Ave. Pine Brook, N.J. Booths 3512-3518

H. R. Foster, ▲ E. E. Crump, John Gilmore, Tom Dougherty, ▲ Stanley Bara, ▲ Irving Sil-berg, ▲ William Hamer, ▲ Stanley Dickstein, ▲ George Murphy, Gerald Becker, ▲ Roy Hueb-



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Attenuators, Sweep Oscillators, Noise Figure Measure, 1 KC- 26,000 MC, *Automatic Noise Figure Measure, Pulsed Carrier Generators, Sound and Vibration Analysis Equipment, Transistor Test Set, 50 MC Cut-Off: Instru-ments For Analysis, Telemetered Signals, Radar IF's and Front Ends. Magna Sweep S-Band Sweep Oscillator Sween Oscillator

(Continued on page 304A)

Elevators at north and east sides of the main lobby take you direct to the Fourth Floor



OPERATION SIMPLIFICATION. A high priority request from the Pentagon points to the big need for simplification in missile control and guidance systems. Fewer components and mechanical parts are what the Defense men are looking for in the hope of cutting down the numerous failures and malfunctions in missile performance. Along the same line, an Army Colonel has called for "deprovement" as a technique for design in military equipment, What he's looking for is stuff that's "smaller and worser" rather than "bigger and better." It's a wide open invitation and everyone is welcome to suggest possible solutions!

ELECTRIC CAR REVIVAL. Fuel cells that generate electricity when hydrogen and oxygen gases are fed into special carbon electrodes may spark a revival of the electric car. Resembling in some respects a big battery, the fuel cell might offer these advantages: silent operation; no smog-producing exhaust; economy because of the high (65% to 80%) efficiency. Some 20 companies are reported working on fuel cells for everything from satellites to outboard motors.

BIG VOICE FOR THE NAVY. By the end of this year the Navy hopes to have the world's most powerful radio station in operation at Cutler, Maine. The antenna system will have two arrays, each having a center tower 980 ft. high; an inner ring of six towers 875 ft. tall; and an outer ring of six towers 800 ft. high. Some 2200 miles of copper will be used in the ground system under the antenna arrays. With transmission output up to two megawatts, radio contact will be possible even with subs lying deep in the Arctic Ocean.

GUIDE TO TESTING. A detailed breakdown of what the Department of Defense demands in the way of environmental testing for electronic component parts is still available. Components fall into one of eight distinct groups and evaluation is done in terms of 12 test areas. Write to the Office of Technical Services, Dept. of Commerce, Washington 25, D.C. and ask for "Environmental Requirements Guide for Electronic Component Parts."

DUGOUT DATA. So far it's been man against man when rival baseball team managers try to outsmart each other. But soon baseball strategy will be decided by a computer. At least that is what's going on at the University of California, where a big computer is being "given its innings." The machine weighs the desirability of thirteen possible plays against the composite batting averages of the typical big league team.

CABLEMAN'S CORNER. The subject of cable testing is an important one. This is the phase of production that determines whether or not the cable you are purchasing is in accordance with your standards and requirements. In the field of electronics and automation, cables are required to suit various stringent electrical, mechanical, and/or chemical environments. Many years of study and testing have gone into the design of test equipment to be used for these critical tests. It is not enough to know that a cable has been tested in a manner that is "essentially" the same as the required standard. Slight variations in equipment design or methods of tests can mean the difference between conformance and non-conformance. Make sure the test data you receive gives a true picture of the performance of your cable. When you need cable, call on a cable specialist. Phone Rome 3000, or write to Rome Cable Division of Alcoa, Dept. 1230, Rome, N. Y.

These news items represent a digest of information found in many of the publica-tions and periodicals of the electronics industry or related industries. They appear in brief here for easy and concentrated reading. Further information on each can be found in the original source material. Sources will be forwarded on request.


I was about to fizzle out . . . the heat really had me! But that McLean Blower did the trick, and now I have a long life ahead.



McLEAN RECESSED MODEL Permanent filter and blower recessed into unused portion of open base rack. Fit 3¹/₂", 5¹/₄", 7", 8³/₄", and 10¹/₂" heights. Deliver 150 to 800 cfm.

Extend the life of sensitive tubes. transistors and other components with McLean packaged cooling units. Prevent system failure . . . maintain calibration and accuracy. Find out how in McLean's 1960 catalog . . . 44 pages of helpful information on cooling electronic equipment. McLean's rack-mounted fans and blowers are smart, compact, easy-to-install and have a multitude of mounting possibilities. Over 100 models in various panel heights and CFM's. Mil.Spec.equipment for packaged cooling also available.





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Whom and What to See at the Radio Engineering Show

(Continued from page 300A)

JFD Electronics Corp. 6101 Sixteenth Ave. Brooklyn 4, N.Y. Booth 1622

▲ Jack Goodman, ▲ Nelson Berman, ▲ Fred Strauss, Barrett Border, Robert Crawford, John Neenan, Richard Moran, Gunnat Hansen, Paul Wohl, Lew Salomon, ▲ Bob Bender, Joseph Murray Development & Manufacture of Precision Metallized Variable Piston Capacitors (Glass & Quartz); LC Toners, Metallized Inductors* (Fixed & Variable); Distributed & Lumped Constant Delay Lines* (Also Tapped & Variable); Filter Networks*, Diplexers*, Multiplexers*, Commercial & Military Antennas and Accessories.

James Electronics, Inc. 4050 North Rockwell St. Chicago 18, Ill. Booth 2100

▲ John A. Kennedy, ▲ Gerald W. Puce, Merwin E. Crowe



James introduces a new line of 'Miniature Choppers, a new Micro-Miniature Transformer Kit, Micro-Scan Sampling Relays, Specialized Input Transformers, Instrument Choppers and Vibrators.

James Vibrapowr Co., Booth 2100 See: James Electronics, Inc.

Japan Electric Industry, Dempa Shinbun, Inc., Booth 2935

2 Kanda Matsuzumi-cho, Chiyoda-ku, Tokyo, Japan &

629 Wellington Ave. Chicago 14, Ill.

Genichiro Oguri, Hideo Hirayama, Hal Hirayama, George Taki Semiconductor Products, Curamic Variable Capacitors, IFTs, Oscillator Coils, Antenna Coils.

Jennings Machine Corp., Booth 4132 3452 Ludlow St. Philadelphia 4, Pa.

T. J. Edwards, J. R. Peace, L. J. Mullin, K. Metz, R. Fleming

Automatic wire measuring, cutting and stripping, and terminal attaching machine.

Jennings Radio Mfg. Corp., Booths 1802-1804

970 McLaughlin Ave. San Jose, Calif.
▲ Jo Enmett Jennings, ▲ Calvin K. Townsend, Robert E. Johnston, ▲ Lewis B. Steward, Don F. Hamm, Raymond B. McGill

Vacuum Fixed & Variable Capacitors, Vacuum Transfer Relays, and Vacuum Power Switches for use in High Powered Transmitters, Power Supplies, and Power Switching Applications,

▲ Indicates IRE member. * Indicates new product. Jerrold Electronics Corp. 15th & Lehigh Philadelphia 32, Pa. Booth 3056 ▲ C. C. Cooley, ▲ K. Simons, ▲ A. Kirkaldy



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Jonathan Mfg. Co., Booth 4514 720 E. Walnut St.

Fullerton, Calif. M. Fritz Hagen, Richard Becker, Charles Meyer

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PROCEEDINGS OF THE IRE March, 1960

World Radio History



William Caprata, Richard Klem, Lloyd Morrow, Harold Sweetser, Robert Bell, Roger Booty, Fied Matthiesen, Joseph Oster Electronic and Cross-har digital to analog conversion, storage and switching. Engi-neering, manufacturing and installation facilities for any type industrial system requiring advanced switching techniques.

Itemco, Inc. 18 Beechwood Ave Port Washington, N.Y. Booth 3053 Albert J. Deeb, ▲ John M. Thompson, Jerome Greenberg, ▲ Socrates Vavoudis, Thomas Loff-man, Edward Migliori, Raymond Dzelzkałus

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J-B-T Instruments, Inc., Booth 2833 133 Hamilton St.

New Haven 8, Conn.

Roland M. Bixler, Louis J. Umile, Carl Iac-carino, Eric Ericson, William Cogger, Charles E. Meyers, William Fitzgerald, William Oliver, Andrew Giannelli, Charles Cole, Mary Gog-gins, Virginia C. Devitt

gins, Virginia C. Devitt Frequency Meters, vibrating reed; Frequency Limit Control; Relays & Oscillator Controls; Toggle Switches, SP, DP, *4-pole types; Switches, lever & rotary selector types; Elapsed Time Meters, including new *2/2," & 3/2," sealed & 400 cycle types; Tachometers; Tempera-ture Indicating Instruments; Electrical Indicat-ing Instruments; Shurite Meters; 'New Edge-Wise Meters.

(Continued on page 302.4)

Your registration admits you to the show for all four days, and to all technical sessions at the Coliseum and the Waldorf-Astoria. Be sure to keep your badge and pocket card with you at all times on the floor. Registrations are not transferable.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Whom and What to See at the Radio Engineering Show

(Continued from page 296.4)

International Rectifier Corp. 1521 E. Grand Ave. El Segundo, Calif. Booths 2901-2903

W. H. Atkinson, J. Conto, ▲ J. T. Cataldo, ▲W. E. Wilson, J. Staluppi, L. Phinney, S. Kramer, A. Scott, D. Conrad, E. Moore, L. Duby, H. Pappas, A. Blumenberg, G. Cafero, J. Oliva, I. Gomora, F. W. Parrish, F. Gift



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Silicon Controlled Rectifiers*—1*, 10*, and 16* Amp-rated series. Complete line silicon zener voltage regulators; *Plug-In Silicon Equivalents to vactum tube rectifiers; wide range silicon power rectifiers rated from 6 to 250 amperes per junction; silicon solar cells & *readout photocells for data handling systems; selenium photocells & sun batteries.

International Resistance Co., Booths 2822-2826

401 N. Broad St.

Philadelphia 8, Pa.

G. Butler, R. Bailey, J. B. Henry, K. Schaefer, R. Rollins, M. Marino, W. Canfield, M. Newbold, D. Johnson, R. Dinsmore, E. Wells, E. Corson, J. Searing, T. Halpern, A. Callanan, P. Troilo.

Complete line fixed and variable resistors—composition, deposited carbon, metal film, power and precision wire wound. Insulated chokes, selenium diodes, and rectifiers. Precision potentiometers, turn indicating dials, etched cable and wiring harness.

International Telephone

& Telegraph Corp. Federal Electric Corporation 621-671 Industrial Ave. Paramus, N.J.



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International Telephone & Telegraph Corp. ITT Components Div. Semiconductor Div. 100 Kingsland Rd. Clifton, N.J. Booths 2510-2520, 2615-2625 Murray Roth, Jerry Wills, Al Ginser, Robert Marquart, Fred Wagner, Ralph DeLuca, John Wood, Arthur Freeman, E. L. Shaw



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Plastic encapsulated silicon rectifiers; Low, medium, high voltage silicon rectifier; High voltage tabular sclenium rectifier; Zener diodes and glass seals.

International Telephone & Telegraph Corp. ITT Components Div., Tube Dept. 100 Kingsland Rd. Clifton, N.J.

Booths 2510-2520, 2615-2625 W. P. Maginnis, W. C. Schmitt, R. W. Deutsch, K. Anspach, C. L. Baxter, J. Wood, C. Malinow, J. Kircher, K. Liddane, K. Pritzleff



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latron storage tubes, Traveling wave tubes, Power tubes and Kuthe hydrogen thyratrons and hydrogen diodes.

International Telephone & Telegraph Corp. ITT Federal Division 100 Kingston Rd. Clifton, N.J. Booths 2510-2520, 2615-2625 J. D. Phelan, E. A. Zappile, Murray Block, A. Isaacs, M. Weiner



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International Telephone & Telegraph Corp. ITT Laboratories 500 Washington Ave. Nutley, N.J.

Booths 2510-2520, 2615-2625 T. Warren, L. Pollack, S. R. Hoh, R. E. Lott, G. E. Howarth, John Burke, E. J. Felesina, J. P. Fitspatrick



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International Telephone & Telegraph Corp. International Electric Corporation Post Office Box 285 Paramus, N.J. Booths 2510-2520, 2615-2625 S. B. Fishbein, A. Anderson, J. Martin, D. Battin



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(Continued on page 300A)



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LABORATORY POTENTIOMETERS—for general DC standards work, meter and thermocouple calibration, precise thermocouple measurements, calibration of resistance networks and other high-accuracy potential measurements.

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The complete line of Rubicon potentiometers, galvanometers, Wheatstone bridges and other instruments is available through any of the Honeywell branch offices throughout the nation.



Whom and What to See at the Radio Engineering Show

(Continued from page 294A)

Inso Electronic Products, Inc., Booth 4054

1200 Commerce Ave.

Union, N.J.

Jerome I. Cohn, H. H. Heyden, Dan Brandon, Vi Howell, Charles J. Kemp, John J. Lightner, Fred W. Link, Allen S. Nace, Homer Pacent, Walter M. Padiak, A. W. Pearsall, Clyde Rush, Frank Welling

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Institute of Radio Engineers I East 79th St. New York 21, N.Y.

Booths in Main Lobby

Subscription, membership and professional group information available. Orders for convention record may be filled out in these booths.

Instron Engineering Corp., Booth 3014 2500 Washington St. Canton, Mass.

E. J. Toll Erb, L. F. H. Flynn J. Tolle, ▲ George S. Burr, ▲ Donald R. L. F. Cedrone, Harold Hindman, Robert

Universal Testing Instruments for study of forcedeflection characteristics of study in thes, sockets, etc. Also used for studying strength char-acteristic of wire, transistor leads and components.

Instrument and Apparatus News, Booth 4307

See: Instruments Publishing Co., Inc.

Instrument Development Labs., Inc., Booth 2106

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Instruments and Control Systems, Booth 4307

See: Instruments Publishing Co., Inc.

First and Second floors-Components Third floor—Instruments and Complete Equipment

Fourth floor-Materials, Services, Machinery

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H. Snort, Max G. Bauer Publications: Instruments and Control Systems, Instruments and Control Systems Buyers' Guide, Instrument & Apparatus News, Military Systems Design, Books: Digital Techniques for Computa-tion & Control, Automatic Control Technology, Printed Circuitry, Nuclear Reactors for Industry & Universities, Mechanical Measurements by Elec-tronic Methods, Strain Gages and others.

International Business Machines Corp. 590 Madison Ave, New York 22, N.Y.

Booths 3507-3511

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Henry Bercz, Beatrice Berez, Charles Kraver, Larry Galvin

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International Electric Corp., Booths 2510-2520, 2615-2625 See: International Felephone & Telegraph Corp.

International Electronic Industries, Inc., Booth 4127 See: Standard Pressed Steel Co.

International Electronic Research Corp. 145 West Magnolia Blvd. Burbank, Calif. Booth 1522 Harvey Riggs, Don Rich, Fred Miller, ▲ John E. Markley, Jr., ▲ John C. McAdam, Ted Rafalovich



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International Electronic Research Corp., ELIN Division, Booth 3018 See: ELIN Division.

International Instruments, Inc., Booth 2813

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See also: F. J. Stokes Corp.

International Radiant Corp. 577 East 156th St. New York 55, N.Y. Booth 3239

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(Continued on page 299A)

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* Indicates new product.

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STABILIZED MICROWAVE SIGNAL GENERATORS

ULTRA STABLE VARIABLE FREQUENCY VERSATILE

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Self-contained

Short-term stability ----

1 part in 10⁸.

10 mw power output for low power applications. Offers many features found only in much higherpriced instruments.



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Whom and What to See at the Radio Engineering Show

(Continued from page 291A)

I.D.E.A., Inc., Booth 1327A See: Industrial Development Engineering Associates.

> IMC Magnetics Corp. 570 Main St. Westbury, L.L., N.Y. Booth 2229

S. Saretzky, J. Wohryzek, G. Egan, A. J. Silverman, A. H. Mankin, L. Seidner, R. W. Kopprasch, C. Burmeister, H. Lamkin, H. Kanter, B. Miller, B. Eder, C. Davis



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ITT Components Div., ITT Federal Div., ITT Industrial Products Div., ITT Laboratories, Booths 2510-2520, 2615-2625 See: International Telephone & Telegraph Corp.

Illinois Condenser Co. 1616 N. Throop St. Chicago 22, Ill. Booth 2310 ▲ J. J. Kurland, A Joseph J. Kurland, L. W.



Motor Running- Motor Starting—Film Capacitors—Mylar, Polystyrene, Filters, Metallized Paper Computer Capacitors—Special Quality Capacitors—Electrolytic Capacitors—Ceramic Encased Paper Capacitors—Ceramic Encased Paper Capacitors—SMT Subminiature Capacitors.

Illinois Tool Works, Licon Switch & Control Div., Booth 1506 See: Licon Switch & Control Div.

Inca Manufacturing Div., Booths 4028-4029

See: Phelps Dodge Copper Products Corp.

Indiana General Corp., Booths 1310-1316 See: Indiana Steel Products Div. & General Ceramics Div.

▲ Indicates TRE member.

* Indicates new product.

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Indiana Steel Products Div. Indiana General Corporation 405 Elm St. Valparaiso, Ind. Booths 1314-1316 P. M. Wheeler, S. C. Beyerl, K. S. Talböt, P. R. Janes, V. B. Farr, J. R. Ireland, R. M. Handren, R. F. Smith, I. A. Dickey, R. W. Moore Complete line of Indox, Alnico V-7*, Alnico VIIA*, Alnico and Cunife Permanent Magnets for the electronics industry. Magnet design consultation service on visitor's problems is offered. Magnetic Materials. Induction Motors Corp., Booth 2229 See: TMC Magnetics Corp.

Induction Motors of California, Booth 2229

See: IMC Magnetics Corp.

Industrial Development Engineering Associates, Inc. 7900 Pendleton Pike Indianapolis 26, Ind. Booth 1327A ▲ E. C. Tudor, ▲ A. C. Elles, C. E. Mathis

Digital and alpha-numeric displays, "New 16 stroke back lighted segmental digital and alpha numeric units, and miniature, high brightness low voltage numeric indicator.

Industrial Electronic Hardware Corp. 109 Prince St. New York 12, N.Y. Booth 2837 Marvin Gottlieb, John Donato, Joel Jacobs Complete Line of *Slide Switches, Molded and Laminated Tube Sockets for Printed

Complete Line of *Stide Switches, Molded and Laminated Tube Sockets for Printed Circuitry, and Conventional Wiring, Transistor Sockets, Connectors, Wired Assemblies, Metal and Bakelite Stampings.

Industrial Hardware Mfg. Co., Inc., Booth 2837

See: Industrial Electronic Hardware Corp.

Industrial Instruments, Inc., Booths 3227-3229

89 Commerce Rd. Cedar Grove, N.J.

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Industrial Products-Danbury Knudsen Div., Amphenol-Borg Electronics Corp., Booths 2501-2503 33 East Franklin

Danbury, Conn. N. Blair, C. Concelman, W. Vockerath, N. Cresci, J. Figueira, I. Pederson, B. Washisko, M. Rose, R. Fowler, J. Wales, J. Huneke RF connectors, including new Crimp MB and microminiature series; Coaxial Relays and switches; Ceaxial Attenuators; Rack & Panel connectors; Custom Wave Guide components. Industrial Test Equipment Co. 55 East 11th St. New York 3, N.Y. Booth 3513 M. Schreibman, ▲ R. Rothschild, C. Laskin, W. Meyer



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Industrial Timer Corporation, Booths 1806-1808

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Industrial Winding Machinery Corp. 120 Woll St., Pox 62 New York 5, N.Y. Booths 1407-4409 A Dr. Henry W. Rochrig, Hans Gamlien



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Industro Transistor Corp., Booth 1628 35-10 36th Ave.

Long Island City 6, N.Y.

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(Continued on page 296.4)

March, 1960

at the IRE SHOW BOOTHS 3301-03-05

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Model 1245 Frequency Range: 1 kc to 300 MC. Measures 5°_{x} at 100 MC. O Multiplier: * 0.9 to = 2. Delta Q: 25:0-25. Test Circuits: separate LF and HF test cir-cuits have ranges of 1 kc to 300 MC. Capacitance Range: 7.5 to 110 µµf with 1-0-1 µµf incre-mental, for either test circuit: 20 to 500 µµf for LF test circuit. Shunt Loss: 12MΩ at 100 MC. External Oscilla-tors : Model 1247. 20 to 300 MC. Model 1246, 40 kc to 50 MC.



Model 1247

Model 1245

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MARCONI INSTRUMENTS for Electronic Measurement

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- * Frequency range : 30 to 300 MC
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- (inductance measured as negative capacitance)* Features high-stability servo-controlled
- conductance balance system This simple, easy-to-use instrument is a general-purpose VHF bridge particularly suitable for measurements on unbalanced antenna systems, coaxial transmission lines, and distributed components in general, as well as on a wide range of lumped components. Operating in the range 30 to 300 MC, it fills the gap between slotted lines and conventional RF bridges. It is arran-

on a wide range of lumped components. Operating in the range 30 to 300 MC, it fills the gap between slotted lines and conventional RF bridges. It is arranged for use with an external oscillator and detector, both of which can be supplied as optional accessories; alternatively a conventional VHF signal generator and receiver may be used.

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Modulation : FM deviation continuously variable and monitored from 0 to 20 and 0 to 100 kc. Also AM up to 40%. Modulation frequencies, 1 and 5 kc. Rack mounting version, Model 1066A/1, now available.

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Whom and What to See at the Radio Engineering Show

Huggins Laboratories, Inc. 999 E. Arques Ave. Sunnyvale, Calif. Booths 2917-2918 A Richard Huggins, A Vern Varenhorst, William Fleig, Bruce Pruitt



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See all the exhibits!

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Mezzanine at back of first floor, South Room at center of south wall, second floor, 3000 court at southeast corner of third floor, 4000 court at southeast corner of fourth floor, 4500 court in northwest corner of fourth floor. Hughes Aircraft Co. Florence & Teale Sts. Culver City, Calif. Booths 1609-1615

Booths 1609-1615 ▲ M. D. Adcock, T. B. Aitkin, T. H. Anderson, W. D. Bird, Jr., R. J. Borstelmann, ▲ E. O. Bowers, J. M. Britton, C. C. Christ, A B. A. Coler, B. W. Davis, J. B. de Groot, H. Dodson, ▲ I. Drukaroff, A. E. Fisher, W. Gould, A. Graydon, ▲ L. A. Gustafson, D. W. Harr, J. D. Hartley, J. P. Hughes, R. T. Jones, J. J. Kowall, J. R. Lyons, R. C. Martens, W. M. Mcc-Hugh, B. O. Moxon, R. T. Orr, ▲ R. B. Reade, F. A. Salvatore, J. J. Stypa, ▲ J. J. Sutherland, J. J. Vogelzang, H. M. Wales, G. D. Wrench



Parametric Amplifier Diode

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High Vacuum Relay Drying, Evacuating and Back Filling System. Completely automatic cycle using dry nitrogen for hermetically sealed relays. Completely automatic Transfer Molding Press for encapsulating small components such as diodes, semiconductors, etc., in latest type epoxy molding powders.

Hycon Eastern, Inc., Booths 3038-3039 See: Hermes Electronics Co.

Hymac Corporation, Booth 3952 See: Molectronics Corp.

> Hysol Corporation Olean, N.Y. Booth 4231

Reginald C. Whitson, A William C. Jenner, Jean Cauchols, Jr., Patrick J. Scutella, Virgil Lorenzini, James A. Collins



Hysol Immediate Flame-Out Flexible Epoxy *HYSOL 15-032, burn proof flexible epoxy casting compound meeting MIL-T27 ASTM and commercial laboratory burn test requirements. *HYSOL 6622 flexible epoxy casting compound family, HYSOL 6700 one component epoxy system, HYSOL fluidized bed and molding powders.

(Continued on page 294A)

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- KNOW that from McMillan they can obtain substantiation to back up performance claims — either in simple, concise form or in advanced, scientific data.
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Whom and What to See at the **Radio Engineering Show**

(Continued from page 288,1)

Hitemp Wires, Inc., Booth 4213 1200 Shames Dr. Westbury, L.I., N.Y.

Walter Merck, Chris Wyer, Frank Lochridge, Robert Martin, William Frogner

High Temperature Wires and Cables. Bondable Teffon Insulated Wires, Ribbon Cables, Teffon Tapes, 1000° Flexible Insulated Magnet Wires.

Hoffman Electron Tube Corp., Booth 1520

See: E.M.L. Cossor Electronics, Ltd.

Hoffman Electronics Corp. Semiconductor Division 1001 North Arden Dr. El Monte, Calif. Booths 1920-1924 G. W. DeSousa, A H. Schoemehl, B. Roberts, A L. Rose, A R. White, A L. Swanson, A W. Prenosil, R. Hoffman, D. Baldwin, H. Freudman, A. R. Clita,

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Honeywell Controls Ltd., Booths 2202-2214 See: Minneapolis-Honeywell,

Hoover Electronics Company 110 W. Timonium Rd. Timonium. Md. Booth 3844 ▲ A. J. W. Novak, ▲ H. K. Schoenwetter, ▲ H. H. Hoge, R. J. McCusker, R. P. Moore, J. B. Miller, G. S. Gay, T. E. Harrowby



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*VERNITEI, high accuracy FM/FM telemeter-ing system components; *Millivolt Transistorized ing system components; *Millivolt Transistorized Subcarrier Oscillator; other transistorized tele-metering products; special airborne electronic in-strumentation; special electronic ground support and control equipment; special electronic test equipment; custom instrumentation vans and malers. trailers

Houghton Laboratories, Inc., Booth 4231 See: Hysol Corporation.

Hovt Electrical Instrument Works, Inc., Booth M-17

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Cambridge 42, Mass.

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Hudson Tool & Die Co., Inc., Booths 4408-4410

18 Malvern St. Newark 5, N. J.

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Complete line of deep drawn metal closures and covers—1800 various standard sizes from sub-miniatures to transformer sizes. Available in MiMetal, Stainless Steel, Aluminum, Brass, Copper, Steel, Monel and others on request.

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Men who know* prefer



a new symbol of magnetic progress



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MAGNETRON Powerful Hyflux ALNICO V magnets improve performance in many types of microwave equipment.



AUTOMATIC DIRECTION FINDER Ferramic "E" magnetic core material helped engineers create a new concept in aircraft antenna design.

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INDIANA GENERAL

PROCEEDINGS OF THE IRE March, 1960

Whom and What to See at the Radio Engineering Show

(Continued from page 286A)

Hermetic Seal Corporation 29-37 South Sixth St. Newark 7, N.J. Booth 1223

▲ M. N. Glickman, ▲ Charles Ward, ▲ Lyle A. Backer, John D. Wood, Al Neumann, Donald E. Hempel, Daniel J. McCarthy, Warren Gressle, Gerald Muller, Joseph Spadafora, W. Speer, H. Perkins, F. Restaino

Perkins, F. Restaino Hermetically sealed microminiatures, miniature and standard seals for relays, frequency control crystals, filters, condensers, rectifiers, transformers, transistors, diodes, Lock-In terminals and glass metal stems. AN connectors, pygmy and "Recon" (rack & panel rectangular) connectors. Gyro headers & terminals, non-magnetic true compression seals. Thermal time delay relays, high vacuum fuses.

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Industry's most complete line of lighted pushbutton switches (Switchlites), Also, complete line indicater lights, toggles, push buttons and basic suap-action switches including environment-free and hermetically-sealed models. Custom electromechanical assemblies include holding coil switches, rotaries, interlocks.

Hewlett-Packard Co. 275 Page Mill Rd. Palo Alto, Calif. Booths 3302-3306. 3101-3405 A David Packard, A William R. Hewlett, A W. Noel Eldred, A B. Oliver, A C. Van Rensselaer, A John Cage, A A. Bagley, A P. Stoft, A N. Schrock, A B. Wholey, A P. Sherrill



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See also: Boonton Radio Corp. & Dymec Division.

Hexacon Electric Co. 161 W. Clay Ave. Roselle Park, N.J. Booth 4012

R. Johnson, J. Grindrod, R. Leary, H. Neff, C. Ellingwood, L. Bryan, W. E. McFadden, J. V. McFadden, R. D. Wood, S. Adlam, R. Dunn, R. Bryan, A. Kulfan, R. Lunneau, J. Enders, R. L. Fish, R. H. Fish, G. Clouse, G. Hamilton, W. Ihlefeld, V. Gummert, A. B. Smith



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Hi-Q Div., Booths 2603-2607 See: Aerovox Corp.

Hickok Electrical Instrument Co., Booths 3616-3618 10514 Dupont Ave. Cleveland 8, Ohio

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PRODUCTION—FOURTH FLOOR

Hill Electronics Inc. 300 N. Chestnut St. Mechanicsburg, Pa. Booth 1903 J. A. Nickerson, A.B. C. Hill, Jr., R. H. Van Zandt, C. D. Bittner



Custom 1 pt 10⁵ HEEMCO Accomplishment

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Carl Hirschmann Co., Inc. 30 Park Ave. Manhasset, L.I., N.Y. Booths 1533-4535 H. Kaefer, J. Bauer, H. Meier, J. Feathe

M. H. Kaefer, J. Bauer, H. Meier, J. Featherstone, A. Grillo, R. Meyer, E. Michow, M. Schrack



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MTCAFIL—Most complete line of coil winding equipment. Small, low cost bobbin winder for laboratories, up to high speed relay winders 18,000 rpm., no causs, no gears to change. Toroidal, precision potentiometer and *fully automatic, high speed armature winding machines.

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Richard M. Livingston. ▲ S. M. Scher, Fred Hough, Irving Becker, Fred Rosenwasser, Richard Hirschmann, Pietre Geervliet, E. Flynn, D. Jaffee, P. Aaron



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(Continued on page 290A)

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960



who needs "for instances"

Take such a "for instance" as this: you want to step an indirectly driven rotary switch immediately on pulse closure. The schematic, Fig. 28, above, points out an easy way of doing it with a standard relay.

Or suppose you'd like to stretch a pulse with a relay – or set up a sequential programming circuit with stepping switches. You'll find the most practical methods among the 41 schematics in AE's new 32-page booklet on "Basic Circuits." It comes complete with a handy template for drawing relays and switches on your own circuits. And it's all yours for the asking ! In case you don't find the solution to your switching problem in *Basic Circuits*, our engineers can find a solution for you. In fact, AE engineers can show you switching tricks that will cut corners on costs. Or, if you wish, they'll assume the job of designing and building to your specs, anything from a simple control package to an entire system.

To get your copy of *Basic Circuits* and/or the answer to your specific control problem, just write J. E. James, Director, Control Equipment Sales, Automatic Electric, Northlake, Ill. In Canada: Automatic Electric Sales (Canada) Ltd., 185 Bartley Dr., Toronto 16, Ont.



GENERAL TELEPHONE & ELECTRONICS



h, 1960

World Radio History

Whom and What to See at the Radio Engineering Show

(Continued from page 284A)

Halliburton, Inc., Manufacturing Div. 4721 S. Boyle Ave. Los Angeles 58, Calif. Booths 4508-4510

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Hanovia Liquid Gold Division, Booths 2110-2118 See: Engelbard Industries Inc.

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Harvard Industries, Inc., Booth 3843B See: Frequency Standards,

Harvey-Wells Electronics, Inc., Booth 3236

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Haveg Industries, Inc., Booth 4418 900 Greenbank Rd. Wilmington 8, Del.

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Haveg Custom Fabricated Silicone Rubber Prod-ucts and Custom Molded Havelex Products nets and Custom Mold (Formerly GE Mycalex).

See also: American Super-Temperature Wires, Inc., Booth 4416

Hayden Publishing Co., Booths 4404-4406

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Gray, Jr.

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Do-it-yourself Heathkit Products, Gm Tube Tester, *TM Test Oscillator, KF Signal Gen-erator, *Transistor Communications Receiver, *Transistor Depth Sounder, *Transistor 3-Band Direction Finder, 6 & 10 Meter Transceivers, *Educational Kit, *Stereo Amplifiers, *Transis-tor Portable Radius, A complete line of Amateur Radio, Test, HI-FI, and Marine Kits.

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Hydraulic-Magnetic circuit breakers including sub-munature, hermetically sealed model. Silic-O-Netic time delay relays, Silic-O-Netic overlaad relays, Adjustable Trans-O-Netic time delay relays.

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(Continued on page 288A)

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 12 db for error rate l in 101: 14 db for NOISE error rate < 1 in 104 600, 1200, and 2400 baud. 1500. 1667, and 2500 baud with internal synchronization or any rate be-tween 500 and 2500 baud with external synchronization. Adjustable from 0.8 to IZATION 3.5 ms: frequency of max, delay settable from 1 to 2 kr.

 WITTER
 +5 volts min., +50 volts max., ground-teferenced digital in-formation at bit rate.

 WITTER
 -20 to +6 dbm
 REQUIRED SIGNAL TO RMS NOISE SPEED 600, 12 DELAY EQUALIZATION TRANSMITTER TRANSMITTER OUTPUT LEVEL -40 to +10 dbm (Au-tomatic Gain Control) •20 volts •10°o, ground-referenced in-formation_at bit rate RECEIVER RECEIVER OUTPUT LEVEL

p,

THE SEBIT-25 is a wire line terminal unit for transmitting and receiving binary information at 500 to 2500 baud (bits/sec) in a nominal 3-kc voice band, such as a long distance toll circuit. This simple AM system (SEBIT-25) uses vestigial sideband transmission and synchronous operation. It includes time delay and amplitude distortion compensating circuits. The equip-ment is 100% transistorized and has been carefully engineered to function properly under a wide variety of environmental conditions. Voice override is included so that the circuit can be used as an order wire. The SEBIT-25 finds use in transmit-ting: high speed data between business machines and com-puters; high speed facsimile information; time division multi-oles information; and tequential transmitting of telemetering data. Write or phone for technical lit-erature, prices, and delivery time.



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(Centinued from page 282.4)

Gudebrod Bros. Silk Co., Inc. 225 W. 34th St. New York 1, N.Y Booth 4025 F. W. Krupp, M. O'Brien, A. Jarnes, J. Paul McDonough, C. C. Schrader, F. Hooven



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Two extra-tough lacing tapes for high-tempera-ture. flat-braided from DuPont TEFLONR, I wo extra-tough lacing tapes for high tempera-ture, flat-braided from DuPont TEFLON R., Temp-Lace II has high fungus resistance; Pre-shrunk Temp-Lace has minimum shrinkage under ex-treme temperature: -40°C, to 220°C.

Gudeman Co., Booth 2129 340 West Huron St. Chicago 10, Ill.

▲ Frank Lakowski, Philip Lovecchio, Edward Glass, J. L. Semple Capacitors, Filters, Pulse Transformers and De lay Lines.



HEEMCO, Booth 1903 See: Hill Electronics, Inc.

H P L Manufacturing Co., Booth 4529 15210 Miles Ave. Cleveland 28, Ohio

Gordon R. Barber, Melvin E. Lorentz, Raynard A. Hedberg, B. Willig, Gilbert Pike, Ronald Pike

Short-run stampings. All materials, 25 to 25,000 pcs. Job shop fabrication.

HRB-Singer, Inc., Booths 1819-1823 See: Singer Manufacturing Co.

Haller, Raymond & Brown, Inc., Booths 1819-1823

See: Singer Manufacturing Co.

(Continued on page 286.4)

CAFETERIA

Second mezzanine. Take elevator 16 from south side of any floor.

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

package-type TWT power amplifiers with NEC's new long life cathode



Production of traveling wave tubes at NEC began seven years ago and introduction of the package-type three years later. As chief supplier to Japan's complex network of microwave communications, NEC has become the world's largest maker of TW tubes. With the high development costs amortized and large manufacturing capacity, NEC is now able to supply these tubes at well below usual prices.

NEC's new doped nickel cathode core material, a 10-year development, increases both emission and tube life. It has been thoroughly field-proven in disc-sealed planar triodes for 2000-mc equipment of a large U.S. systems manufacturer (name on request). With its cooler operating temperature, evaporation rate of oxide is less than any other known core materials. This extends tube life up to 50%.

Designers will appreciate the compactness these tubes will give to their systems and operators the reliability and economy. Tubes connect to standard IEC waveguide flanges and can be shipped from stock.



4W76

The 4W76 operates in the 4000-mc band and has nominal saturated power output of 10 watts. High amplification over a wide range of power levels results in small-signal gain of approx. 30 db. The band width at half-power points is 1400 mc, but the tube can be used in the frequency range of 2800 to 5000 mc.

Typical Operating characteristics at 4000 in	Typical	Operating	Characteristics	at	4000	mc
--	---------	-----------	-----------------	----	------	----

First Anode Voltage	2,640 V	Saturated Power Output	12.5 watts
Helix Voltage	3,220 V	Small-Signal Gain	32 db
Helix Current	0.7 mA	Noise Figure appr	ox. 25 db
Collector Current	33 mA	VSWR less than	2 to 1
Focusing Electrode Volt	age -40 V	(from 3500	to 4300 mc)

NEC TRAVELING-WAVE AMPLIFIERS

PERM/ 4W75 4W76 6W50	NENT MAGNE 4000-mc band 6000-mc band	T FOCUSED 1.5 watts 5-10 watts 5-10 watts	AMPLIFIERS 8W75 70 8W76 11W17 110	S DOO-mc band DOO-mc band	1.5 watts 5·10 watts 1.0 watt
ELECT 4W85 4W86	ROMAGNET FC 4000-mc band	OCUSED AMP 0.1 watt 1.0 watt	LIFIERS 4W72A 40 7W52 60	000-mc band 000-mc band	1.5 watts 5-10 watts

Advantages of package-type

NO focusing or impedance matching at installation

NO dummy space for removal

NO power source or current stabilizer for electromagnet



Whom and What to See at the **Radio Engineering Show**

(Continued from page 279A)

Gorman Machine Corp. 480 S. Main St. Randolph, Mass. Booth 4032 Kenneth P. Gorman, Harvey R. Snider, Mona I. Collum



New and Versatile High Speed Subminiature Toroid Coil Winding Machine Model #4000, Capable of Winding Coils As Small As .055 1.D. To As Large As 2" O.D. Without Changing Heads In Operation at Booth.

Gould-National Batteries, Inc., Booth 1116

See: Nicad Division.

Granite State Machine Co., Inc., Booth M-25

124 Joliette St.

Manchester, N.H. ▲ Joseph A. Cassidy, ▲ James A. Banker, George Tack

Antennas, antenna control systems, antenna tun-ing systems, multicouplers, radar target simula-tors, servo-mechanisms, gears and gear trams, sheet-metal fabrications, technical publications.

Grant Pulley & Hardware Corp. 43 High St. West Nyack, N.Y. Booth 4112

N. A. Gussack, Millon Gussack, William M. Linden, Jerry Bross, Lou Wichers, H. Schaef-fer, Jack Vissman, Charles Agnoff, Walter Hess, Al Nelson, Joseph Norton, Robert Saun-ders, Robert Lawson, Alvin Ring.





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Graphik Circuits Div., Booth 2535 See: Cinch Mfg, Company,

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Greibach Instruments Corp. 315 North Ave. New Rochelle, N.Y. Booth 3924

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(Continued on page 284A)



WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960



This model of the binary symmetric channel symbolizes the probability of error, p. How can a one received in the zero slot be caught and corrected?

Group codes for prescribed error patterns

Information signals, representing zeros and ones, are transmitted through a binary symmetric channel at such high speeds that they are subject to channel noise. Through group codes it is possible to detect and correct automatically large classes of errors that may arise from such disturbances.

Usually, in optimizing these codes all possibilities are classified and samples of each are evaluated. But this task can become enormously complex. For large information blocks, such as a 70-place code, there may be billions of possibilities to

evaluate. To reduce the need for these ex- rected, codes with a minimum number of haustive methods, IBM scientists have evolved a preliminary theory for constructing group codes through a correlation analvsis of error patterns.

Correlated patterns of errors are organized into equivalent classes and a code is formulated to overcome the error-producing characteristics of the communications channel. A code for one pattern of errors may be transformed mathematically into codes for other patterns of the same class. By prescribing which error patterns can be corchecking signals may be formulated.

This optimizing process can have practical significance since every checking signal for a given number of information signals in a group code increases the cost and delay in information processing. In addition to the work described here, other approaches to the problem of code simplification are being made at IBM through linear programming and computer simulation,

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World Radio History

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(Avtomatika i Telemekhanika)	Monthly	Abstracts only		Office of Technical Services U. S. Dept. of Commerce Washington 25, D. C.
Journal of Abstracts, Electrical Engineering (Reserativnyy Zhurnal: Electronika)	Monthly	Abstracts of Russian and non- Russian literature		Office of Technical Services U. S. Dept. of Commerce Washington 25, D. C.
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World Radio History

Electroluminescent Cell Applications-Lochinger and Strutt, (See 442.)

621.385:534.29

430

Microphony of Thermionic Valves-A. Stecker. (Elektronik, vol. 7, pp. 361-367; December, 1958.) Examples of stroboscopic analysis are given. See also 1013 of 1959.

621.385 032 213 13

The Study of Interface Layer of Oxide-Coated Cathode by X-Ray Diffraction Method -T. Imai and II. Niizeki. (Rep. elect. Commun. Lab., Japan, vol. 7, pp. 55-70; March, 1959.)

621.385.3.004.6:621.395.64

The Life of Wide-band Amplifier Tubes-T. Kojima and H. Hara. (Rep. elect. Commun. Lab., Japan, vol." 7, pp. 36-43; February, 1959.) Factors which control the life of tubes for coaxial-cable repeaters are given.

621.385.61

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Amplification of Space-Charge Waves in the Movement of a Beam of Electrons with Variable Velocity through Channels in a Medium with Ohmic Losses-Yu. F. Filippov. (Radiotekh. Elektron., vol. 4, pp. 228-232; February, 1959.) Theoretical investigation of the propagation of a modulated beam with variable velocity in a medium with ohmic losses. Expressions are derived for a particular case of exponentially increasing waves.

621.385.6

The Motion of a Beam of Electrons Moving with Periodically Varying Velocity through the Channels of a Medium with Ohmic Losses-Yu. F. Filippov. (Radiotekh. Elektron., vol. 4, pp. 233-240; February, 1959.) Theoretical investigation of the phenomenon occurring when an electron beam moves through a porous medium with ohmic losses. Contrasting with the case of propagation of an electron beam in vacuum, exponentially increasing waves exist for all values of the parameters of the beam and the medium. It is shown that the presence of thermal velocity does not lead to a critical frequency.

621.385.6:621.3.032.269.1

Experimental X-Band Preamplifier Tubes with 4.5 db Noise Figure-J. E. Nevins and M. R. Currie. (PRoc. IRE, vol. 47, pp. 2015-2016; November, 1959.) The low noise is achieved by improving electron gun design.

621.385.623.5

Experimental Investigation of the Synchronous Operation of Reflex Klystrons in the Three-Centimetre Band-S. D. Gbozdover, A. I. Kostienko and G. P. Lyubimov. (Radiotekh. Elektron., vol. 3, pp. 105-111; January, 1958.) The paired-series synchronous operation of several reflex klystrons operating into a common load with displaced frequency characteristics is examined. It is shown that this operation can be accomplished without discontinuities in the frequency spectrum or in the generated power. The tuning range for systems of several klystrons may exceed the sum of the separate tuning ranges.

621.385.623.5

The Effect of a Load on the Synchronous Operation of Two Reflex Koystrons-A. I. Kostienko and G. P. Lyubimov. (Radiotekh. Elektron., vol. 3, pp. 112-115; January, 1958.) Experimental investigation of the dependence of the bandwidth of synchronization and the tuning range of two synchronously operating klystrons on the impedance/frequency characteristic of the load in the 3-cm range.

621	.385	.623.5:62	1.317.34				7	07
	The	Spectral	Density	of	the	A.M.	Noise	in

Reflex Klystrons-H. Häggblom. (Proc. IEE, Part B, vol. 106, pp. 497-500; November, 1959.) The AM noise spectrum is treated experimentally and theoretically for signal frequencies of 4.7 and 9.3 kmc. By considering the synchronizing effect of the RF components in the electron beam, the theory gives a nearly constant noise spectrum within a relative bandwidth of about $10^{-4}-10^{-3}$.

621.385.624

Triple-Resonator Klystron Frequency Multipliers-A. D. Sushkov. (Radiotekh. Elektron., vol. 4, pp. 246-252; February, 1959.) Description of a triple-resonator klystron with prelitninary signal amplification and with different frequency multiplication factors. Theoretical and experimental results are examined. Comparison of the performance of triple-resonator and double-resonator types shows the superiority of the former.

621.385.624

Frequency Conversion with a Reflex Klystron-E. N. Bazarov and M. E. Zhabotinskil. (Radiotekh. Elektron., vol. 4, pp. 253-261; February, 1959.) Mathematical analysis indicating theoretically the possibility of using a reflex klystron for frequency division and multiplication. Five operating modes are considered and a new reflex klystron is proposed which can increase the coefficient of electron interaction during processes of frequency conversion. This klystron consists of two cavity resonators with three or four grids.

621.385.624 710 Some Experiments on Broad-Band C.W. Power Klystrons at X Band-M. O. Bryant and C. P. Lea-Wilson. (J. Electronics Control, vol. 6, pp. 481-498; June, 1959.) The experiments were designed to assess the handwidth capability of a multicavity klystron for CW powers of several hundred watts. This is shown to depend on maximum electron current density and hence on magnetic focusing field. Its gain-bandwidth product is better than that of a traveling-wave tube, but not its efficiencybandwidth product.

621.385.63

711 Computing the Power of Interaction between an Electron Beam and the Field of a Delay System with a Given Field Approximation-V. M. Lopukhin and G. A. Sitnikova. (Radiotekh. Elektron., vol. 4, pp. 218-227; February, 1959.) Solution of the problem of the excitation of a slow-wave structure in which the field consists of a superposition of n harmonics with amplitudes that increase with the coordinates. Expressions are derived for the current density, the power of interaction between the electron beam and the field, and efficiency. A solution is obtained for the excitation of a traveling-wave tube with a variable average electron velocity; the effect of a variation in the average electron velocity on the excitation conditions for the system is examined. The effect of the space charge on the power of interaction between the electron beam and the field in an exponentially increasing wave is considered.

621.385.63

An Isochronous Travelling-Wave Valve-G. F. Filimonov. (Radiotekh. Elektron., vol. 3, pp. 85-93; January, 1958.) The investigation shows the possibility of increasing by 3 db the high-frequency field of a traveling-wave tube. The effect of a number of parameters on the magnitude of this effect is also examined.

621.385.63

Nonlinear Theory of a Travelling-Wave Valve: Part 3-The Effect of the Repulsion Forces-L. A. Vainshtein and G. F. Filimonov. (Radiotekh. Elektron., vol. 3, pp. 80-84; January, 1958.) An investigation of the effect of Coulomb repulsion forces on the operation of a traveling-wave tube amplifier. Nonlinear equations are used and results are compared with those obtained by other authors.

Part 1-2426 of 1959.

Part 2-vol. 2, pp. 1027-1047; August, 1957

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An Electron Beam in a Helix Placed in a Dielectric Medium-V. P. Shestopalov. (Radiotekh. Elektron., vol. 3, pp. 131-141; January, 1958.) Investigation of em wave propagation in a helix located in a dielectric medium in the presence of an electron beam. Equations for the em field and the electron motion are derived. The dispersion equation is also examined and a graphical solution of this is shown.

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Travelling-Wave Tube Equations including the Effects of Parametric Pumping-I. S. Cook and W. H. Louisell. (PRoc. IRE, vol. 47, p. 2016; November, 1959.) A general set of equations is given for investigating the process of parametric amplification of space-charge waves in traveling-wave-tube electron beams. Other data on the derivation and application of the equations are to be published later.

621.385.632

Travelling-Wave Valve Oscillator with an Electronic Phase Shifter-M. V. Tychinskii and V. G. Fedorov. (Radiotekh. Elektron, vol. 4, pp. 241-245; February, 1959.) Investigation of a traveling-wave tube oscillator with an electronic phase shifter located between the helix segments, the phase shifter being a drift tube with a controlled potential. Several oscillation ranges were observed in the frequency band near 3 kmc. The range of electronic tuning reached 4 per cent.

621.385.632.3

On the Theory of the Electron Wave Tube with Elliptic Cross-Section-P. Mattila. (Acta Polytech., no. 241, EL1, 78 pp.; 1958.) Theory is derived and a comparison made with the electron wave tube of circular cross section.

621.385.633

Experimental Investigation of a Nonretarded Backward-Wave Oscillator-K. Ya. Lizhdvol. (Radiotekh. Elektron., vol. 4, pp. 212-217; February, 1959.) Results of an experimental investigation of an UHF oscillator using the interaction between an electron beam and an em wave having a phase velocity equal to the velocity of light. Graphs show the dependence of the wavelength on the flux density of the magnetic field and on the voltage of the electric field. A possible method of frequency modulation is examined.

621.385.64

A New Analysis of Magnetron-D. Kobayashi. (Rep. elect. Commun. Lab., Japan, vol. 7, pp. 100-115; April, 1959.) A general analysis which allows the calculation of the frequency of oscillation as a function of anode voltage and magnetic field.

621.385.64

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Experimental Investigation of the Electronic Conductivity of the Space-Charge Cloud in a Magnetron-V. P. Tychinskii. (Radiotekh. Elektron., vol. 3, pp. 116-130; January, 1958.) A description of a method for measuring the electronic conductivity and the cyclotron resonance cuives in magnetrons. In several cases a satisfactory agreement was obtained between experimental and theoretical results.

621.387

Fluctuation of Starting Voltage in Gas-Discharge Tubes due to Statistical Time-Lag-T Dote. (Rep. elect. Commun. Lab., Japan, vol. 7, pp. 116-122; April, 1959.)

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saturation and line information even in linear transmission channels is considered. Studio circuitry for signal switching and distribution is investigated to assess distortion and crosstalk effects.

621.397.2:621.3.018.782 675 Waveform Distortion in Television Links: Part 2—The Measurement and Correction of Waveform Distortion—I. F. Macdiarmid. (P.O. elec. Engrg. J., Part 3, vol. 52, pp. 188–195; October, 1959.) The techniques of waveform measurement and distortion specification are discussed. An "echo waveform corrector" is described in which suitably attenuated inverted replicas of a distorted waveform delayed or advanced in time are added to the distorted waveform to correct it. Part 1—4209 of 1959.

621.397.2:621.396.665 A New Approach to Balanced Audio Levels in Television—R. B. Montoe. (J. Soc. Mol. Pict. Telev. Engrs., vol. 68, pp. 538-541; August, 1959.) In the system described AGC amplifiers with modified gain characteristics are used on conjunction with manual gain controls to balance AF levels.

621.397.2:681.84.087.7 677 Stereophonic TV Sound—(*Electronics*, vol. 32, p. 64; October 30, 1959.) A description of a multiplex system in which two AF channels are modulated at the horizontal synchronization-pulse frequency and applied to a conventional sound transmitter.

621.397.23 678 Synthetic Highs—an Experimental TV Bandwidth Reduction System—W. F. Schreiber, C. F. Knapp and N. D. Kay, (J. Soc. Mot. Pict. Telev. Engrs., vol. 68, pp. 525–537; August, 1959. Discussion.) A complete system is described which codes a standard video signal to match a channel of narrower bandwidth and subsequently decodes the received signal for display on a standard television monitor.

621.397.6.001.4:535.623 679 Application and Future Development of the Test-Line Technique—II. Springer. (*Rund-funktech. Mitt.*, vol. 3, pp. 40–50; February, 1959.) Experimental transmissions of test lines during the vertical scanning interval by West German transmitters [see 587 of 1956 (Fröling)] were used to investigate white-level functions and to develop automatic white-level functions equipment which is described. For English version see *E.B.U. Rev.*, no. 53A, pp. 2–11; February, 1959. Other papers on test-line technique are given in 1957 IRE NATIONAL CONVENTION RECORD, vol. 5, pt. 7, pp. 17–50.

621.397.61

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The Television Transmitter Ochsenkopf— (*Rundfunktech. Mitt.*, vol. 3, pp. 2–28; February, 1959.) A group of papers dealing with the new high-power transmitter in North Bavaria: a) **The State of Television Coverage** in

Bavaria after the Opening of the High-Power Television Transmitter Installation Ochsenkopf-F. M. Daser (pp. 2-4).

b) Television Covarage in the Region of the Ochsenkopf Transmitter—E $Graff~(pp,~5\mathcharceros),$

c) Technical and Economic Problems of the Ochsenkopf Transmitter Installation—E Kessler (pp. 8-11).

d) Load Specification and the Constructional Solution adopted for the Television Tower on Ochsenkopf—F. Staiger (pp. 11–14).

e) Planning and Construction of the Television Tower on Ochsenkopf—G. Jauch (pp. 15-18).

f) Technical Equipment of the Television Tower on Ochsenkopf—E. Angermüller (pp. 19–28).

For a description in English of the television tower see *Engineer* (London), vol. 207, pp. 863– 864 and 901; May 29, and June 5, 1959. 681

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621.397.61

High-Power Television Transmitters for Bands IV and V—T. S. Robson and T. M. J. Jaskolski. (*Proc. 1EE*, Part B, vol. 106, pp. 528–540; November, 1959.) The power requirements of transmitters are considered in relation to the limited coverage obtainable at UHF. Details are given of the Crystal Palace transmitter for propagation tests covering the 405and 625-line systems.

621.397.61.029.63

High-Power Television Transmitter for Bands IV/V (Haardtkopf)—A. Kolarz and A. Schweisthal. (Rundfunktech. Mitt., vol. 3, pp. 29-39; February, 1959.) Description of a complete transmitter installation in southwest Germany with high-power klystrons in the output stages.

621.397.62:616-001.26 X-Ray Emission from Television Sets— C. B. Braestrup and R. T. Mooney. (*Science*, vol. 130, pp. 1071-1074; October 23, 1959.) An estimate is made of the average radiation dose to the gonads from home television sets based on measurements on typical television tubes.

621.397.743 684 Gap-Filling Translators and Transmitters— W. J. Morcom. (J. Brit. IRE, vol. 19, pp. 649– 659; October, 1959.) A general description of the system and apparatus used in low-power television booster stations with details of the performance specifications.

621.391.81:621.397 685 The Field Strengths required for the Reception of Television in Bands I, III, IV and V —G. F. Swann. (*Proc. IEE*, Part B, vol. 106, pp. 541–544; November, 1959. Discussion, pp. 545–547.) The characteristics of the receiving installation are considered together with the statistical variation of field strength with antenna location. The nominal limit of the service area in terms of median field strength is then derived.

TUBES AND THERMIONICS

621.382.2 686 Measurement of Thermal Behaviour of Semiconductor Diodes—(). Jakits. (*Elektrotech. Z.*, vol. 80, pp. 518–520; August 1, 1959.) Measurements of thermal inertia are interpreted on the basis of a simplified model. See also 3109 of 1959.

621.382.2:546.681'18 687 Electrical Characteristics of some Gallium Phosphide Devices—J. Mandelkorn. (PRoc. IRE, vol. 47, pp. 2012–2013; November, 1959.) Characteristic curves of GaP diodes are given; these devices are potentially useful for high temperature work and in rapid switching. A few showed curves very similar to those of a Ge point-contact transistor.

621.382.2:621.317.3

The Measurement of Semiconductor Diode Switching Characteristics—Barry and Fisher. (See 601.)

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621.382.2:021.375.9 689 Semiconductor Varactors using Surface Space-Charge Layers—W. G. Pfann and C. G. B. Garrett. (Pkoc. IRE, vol. 47, pp. 2011-2012; November, 1959.) Such devices could be made by bringing together an *n*-type semiconductor and a nonconducting oxide. They should compare favorably with *p*-*n* diodes for parametric amplifiers both in respect of cutoff frequency and capacitance voltage characteristic.

621.382.2:621.375.9 690 Alloyed, Thin-Base Diode Capacitors for Parametric Amplification—Mortenson. (See 441.) 621.382.23 691 The Temperature Dependence of the Electrical Characteristics of the Silicon Alloyed Junction—II. Izumi, (*Rep. elect. Commun. Lab., Japan,* vol. 7, pp. 123–132; April, 1959.) A report of observations made on pointedjunction diodes in the temperature range 20°K-373°K.

621.382.23 The Time Lag of *p*-n Junction Diodes Strongly Driven with Sinusoidal Voltages—W. Heinlein. (*Arch. elekt. Übertragung*, vol. 12, pp. 510–514; November, 1958.) The theory given is confirmed by the results of tests on Ge diodes driven at different frequencies. See also 2767 of 1959.

621.382.23 693 The Tunnel Diode—Its Action and Properties—B. Sklar. (*Electronics*, vol. 32, pp. 54–57; November 6, 1959.) Electron energy-band diagrams are used to explain the negative conductance property of a semiconductor with high impurity concentration. Breakdown characteristics are illustrated and the equivalent circuit of a tunnel-diode amplifier is given.

621.382.3 694 Special Semiconductor Issue—(*Elektrotech. Z.*, vol. 80; August 1, 1959.) The following papers, which are mainly reviews of transistor applications, particularly in German equipment, are included in this issue; others are abstracted separately.

a) Semiconductor Components, their Physical Properties and Technical Development—II. J. Thuy and R. Wiesner (pp. 473-480), 115 references.

b) Application of Semiconductor Components in Radio and Television Engineering— E. Ginsberg (pp. 481-483).

c) The Transistor in Telecommunications
by Wire—T. F. Grewe (pp. 483-487).
d) Transistors in Control Systems with

d) Transistors in Control Systems with Logical Circuit Elements—E. Götz, H. C. Heinzerling and H. G. Lott (pp. 487-492).

e) Semiconductors in Telecontrol—A. de Quervain (pp. 492-495).

f) Transistors in Digital Measurement Techniques—B. Rall (pp. 495–498).

g) Transistor Switching Circuits in Automatic Computers—II, Weber (pp. 498-502).

h) The Silicon Rectifier in Converter Techniques—E. Nitsche and F. Pokorny (pp. 506-512).

i) Application of Power Rectifiers in Charging Equipment—J. Balkow (pp. 512-514).

j) Chopper Circuits with Transistors—E. Gelder (pp. 520-522).

621.382.3 695 The Physical Interpretation of Measurements on Transistors—S. Deb. and A. N. Daw. (*J. Electronics Control*, vol. 6, pp. 552–553; June, 1959.) Comment on a criticism by Hyde (*J. Electronics Control*, vol. 6, pp. 362–364; April, 1959) of an earlier paper by the authors (2043 of 1959).

621.382.3:621.318.57 The Electrical Properties of Storage-Type Switching Transistors—W. v. Münch. (Nachrichtentech. Z., vol. 11, pp. 565–571; November, 1958.) The characteristics of the transistor, described in 2777 of 1959 (v. Münch and Salow), are derived and discussed.

621.383:535.215 697 Photoresistors, Photodiodes and Phototransistors: Properties and Characteristic Data —P. Görlich, A. Krohs and W. Lang. (*Arch. tech. Messen*, nos. 274 and 275, pp. 235–238 and 247–250; November and December, 1958.) Detuils are given of a number of commercial types of international manufacture. See also 3520 of 1959.

N. F. Barber and D. D. Crombie, (J. Atmos. Terr. Phys., vol. 16, pp. 37-45; October, 1959.) The reflection coefficient is calculated for a sharply bounded ionosphere under conditions applicable to propagation around the magnetic equator. The reflection coefficient depends on whether the direction of propagation is from east to west or vice versa. Possible ways in which direct experimental evidence could be obtained are outlined.

621.391.812.63.029.51

Polarization Computations by means of the

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Multislab Approximation—A. J. Ferraro and J. J. Gibbons. (J. Atmos. Terr. Phys., vol. 16, pp. 136-144; October, 1959.) Calculations have been made by a method requiring digital computer techniques for determining the polarization of LF echoes due to reflection at vertical incidence. A comparison with the more rapid step method of Becker (e.g., 672 of 1952) shows that this latter approach yields satisfactory results

621.391.812.63.029.62

Evidence for a 200-Megacycles-per-Second Ionospheric Forward-Scatter Mode associated with the Earth's Magnetic Field-J. L. Heritage, S. Weisbrod and W. J. Fay. (J. Geophys. Res., vol. 64, pp. 1235-1241; September, 1959.) Signals from a pulsed transmitter in southern Texas were received in California and Nevada in the vicinity of magnetic specular contours calculated for an assumed 100 km height. Rapid fading and broad azimuthal distribution were characteristic features and, although burst-like behavior was most common. steady signals were sometimes observed for 30minute periods. No signals were received from the great-circle direction or north of it. Evidence of the nature and reliability of the scattering is presented.

621.391.812.8

The Effect of the Earth's Magnetic Field on M.U.F. Calculations—K. Davies. (J. Atmos. Terr. Phys., vol. 16, pp. 187-189; October, 1959.) The effect of neglecting the earth's field in the deduction of MUF factor from a vertical-incidence ionogram is to reduce the factor to about 5 per cent below its correct value.

621.391.812.8

How Many M.U.F.s?-T. W. Bennington. (Wireless World, vol. 65, pp. 537-538; December, 1959.) Discussion of the term MUF and the various interpretations and values assigned to it in different circumstances, with particular reference to the recent CCIR assembly.

RECEPTION

621.391.81

The Propagation Conditions between Europe and North America in the 40-52-Mc/s Waveband during the Sunspot Maximum-H. Wisbar. (Nachrichtentech. Z., vol. 11, pp. 586-590; November, 1958.) Data on the reception of U.S. amateur transmissions in the 5-meter band and of telephony transmission at 43.6 mc are analyzed with reference to f_0F_2 and geomagnetic conditions for periods between November 21, 1957, and March 8, 1958.

621.391.812.3

Some Remarks on the Analysis of Fading in the m- and dm-Range-J. Grosskopf. (Nachrichtentech, Z., vol. 11, pp. 577-586; November, 1958.) Correlation methods of analysis and their application to VIIF propagation tests are examined. Scatter-propagation theories are discussed with reference to measured correlation functions.

621.391.82:621.315.62

Radio Interference from H.V. Insulators-C. H. W. Clark. [Elec. Rev. (London), vol. 165, pp. 491-497; October 23, 1959.] The mechanism of interference generation by HV insulators is explained and the effect of insulator design is discussed.

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621.396.62:621.317.794

The Sensitivity of a Radiometer with A.G.C .- N. V. Karlov. (Radiotekh. Elektron., vol. 3, pp. 74-79; January, 1958.) Expressions are obtained for the sensitivity of a modulationtype radiometer in which the sensitivity threshold is controlled by means of an automatic gain control. It is shown that in practical applications this natural sensitivity threshold does not vary considerably.

621.396.62:621.376.3 658 A Quality-Checking Receiver for VHF FM Sound Broadcasting-C. G. Mayo and R. E. Jones. (B.B.C. Engrg. Div. Monographs, no. 25, 15 pp.: June, 1959.) Two receivers with the same basic design are described and the results of standard performance tests are given. In the final model, a combined limiter and discriminator circuit is used [1379 of 1958 (Mayo and Head) |.

621.396.62:621.391.812.7

VHF/FM Multipath-Propagation Test Set -(Wireless World, vol. 65, pp. 559-562; December, 1959.) Description of the effects of multipath propagation on VHF FM signals and a full description of a test set developed to enable this type of propagation to be investigated and its effects evaluated.

621.396.66:621.372.56.029.6

Pre-Receiver Attenuator reduces Intermodulation-S. L. Robinette. (Electronics, vol. 32, pp. 64-65; November 6, 1959.) Brief details are given of a coaxial and a transformer type of magnetically controlled ferrite attenuator for UHF.

STATIONS AND COMMUNICATION SYSTEMS

621.391:621.376.5

Principle of Quantization of Stochastic Signals with Infinite Spectrum and some Results of the Theory of Pulse Transmission of Information-N. A. Zheleznov. (Radiotekh. Elektron., vol. 3, pp. 3-18; January, 1958.) With pulse modulation, a potential fidelity in reception exists which cannot be exceeded with any form of modulation. A method is described for the selection of transmission-line characteristics guaranteeing this degree of accuracy.

621.396:681.84.087.7

662 A Method for the Stereophonic Transmission of Broadcasts-H. Jubisch and H. Seidel. (Elektron. Rundschau, vol. 12, pp. 377-382; November, 1958.) A system of normal bandwidth using amplitude and frequency modulation of a single carrier in the VIIF band is described. Phase inversion is used to eliminate crosstalk.

621.396.2:523.5

Tests made on a Meteor-Burst VHF System-(Brit. Commun. Electronics, vol. 6, p. 781; November, 1959.) A three-year trial of the Canadian "Janet" intermittent communication system [910 of 1958 (Davis, et al.)] is reviewed. Stations 800 miles apart operated 2-kw transmitters at 49 mc using 5-element Yagi antennas and magnetic-tape storage. With a transmission speed during bursts of 2400 wpm, an average speed of 40 wpm with character error rate 0.35 per cent was obtained. Interference sources are noted.

621.396.41:551.507.362.2

Multiplexing Techniques for Satellite Applications-O. B. King. (Electronics, vol. 32, pp. 58-62; October 30, 1959.) A detailed account is given of the ten-channel transistorized timedivision multiplex system used in the satellite Explorer VII, Particular reference is made to

the gating technique, and to the means of analog conversion and storage of random input signals.

621.396.43:551.501.362.2 665

Satellite Communication-C. T. McCov: R. Pierce and R. Kompfner. (PRoc. IRE, vol. 47, pp. 2019-2020; November, 1959.) Comment on 2012 of 1959 with authors' reply.

SUBSIDIARY APPARATUS

621.3.087.9:621.374 666 Using Digital Techniques in Time Encoders-R. J. Sullivan, I. Eastman and I. C. Chanock. (Electronics, vol. 32, pp. 80-83; November 13, 1959.) A decimal indication of the elapsed time is given every 20 seconds. The output is suitable for either magnetic tape or paper chart recorders.

621.314.1:621.372.52 667 The Design of Transistor Push-Pull D.C.

Converters-W. L. Stephenson, L. P. Morgan and T. H. Brown. (Electronic Engrg., vol. 31, pp. 585-589; October, 1959.) The circuit least affected by external influences is a squarewave oscillator controlled by a saturating transformer. Design formulas are given.

621.314.1:621.382.3

Equations for Designing Transistor Power Supplies-T. Hamm, Jr. (Electronics, vol. 32, pp. 122-124; October 23, 1959.) Basic design equations and graphs for transistor dc/dc converters are given.

621.314.63:546.28

660 Operational and Storage Life of Silicon Rectifiers-C. L. Hanks. (Electronics, vol. 32, pp. 82-84; October 16, 1959.) Results are given of life-test measurements of forward voltage drop and reverse current.

621.314.63:621.311.6 670

Controlled Rectifiers drive A.C. and D. C. Motors-W. R. Seegmiller. (*Electronics*, vol. 32, pp. 73-75; November 13, 1959.) By using saturable magnetic-core firing circuits, the size and weight of switching devices are reduced. Half-wave and full-wave push-pull circuits are described

621.314.63:621.382.2 671

Contribution to the Problem of Limiting Values for Semiconductor Circuit Elements-H. Carl and H. L. Rath. (Elektrotech. Z., vol. 80, pp. 502-506; August 1, 1959.) The danger of exceeding the permissible voltage limits in series-connected power-rectifier elements is discussed. This difficulty can be overcome by the use of parallel resistors or capacitors as shown in the reverse-voltage/time curves given.

621.316.721/.722:621.382.3 672

Voltage and Current Stabilization with Power Transistors-W. Müller-Warmuth. (Z. angew. Phys., vol. 10, pp. 497-499; November, 1958.) Basic circuits and design data and details of two power units are given.

TELEVISION AND PHOTOTELEGRAPHY 621.397.132

N.T.S.C. Colour-Television Signals-J. Davidse. (Electronic Radio Eng., vol. 36, pp. 416-419; November, 1959.) Continuation of 335 of January. Experimental results are presented and theoretical consideration is given to the effects of prolonged high subcarrier levels and their influence on the luminance, and to the distribution of the momentary level of the luminance signal.

621.397.132

Transmission Faults in N.T.S.C. Channels H. Schönfelder. (Arch. elekt. Übertragung, vol. 12, pp. 497-509; November, 1958.) The possibility of cross-modulation between luminance and chrominance signals and between

potential Electron Lenses on the Basis of Measurements of the Emergent-Ray Tangents ----K. J. Hanszen. (Z. Naturforsch., vol. 13a, pp. 409–414; May, 1958.)

621.385.833 624 The Influence of Inelastically Scattered Electrons on the Contrast of Plane Objects in Electron Microscopes—W. Lippert. (Z. Naturforsch., vol. 13a, pp. 274–278; April, 1958.)

621.387.4

Gas Čerenkov Counters—J. H. Atkinson and V. Perez-Mendez. (*Rev. Sci. Instr.*, vol. 30, pp. 864–868; October, 1959.)

621.387.4:621.383.27:535.37

Luminescent Effects in Photomultiplier Tube Faces and Plexiglas Čerenkov Detectors --K. A. Anderson. (*Rev. Sci. Instr.*, vol. 30, pp. 869-873; October, 1959.)

621.396.662:[615.8+621.365.5 627 Automatic Tuning of Medical and Industrial High-Frequency Generators—L. Rausch. (Elektronik. vol. 7, pp. 335–336; November, 1958.) A motor-driven tuner unit controlled by a phase-discriminator circuit is described.

621.397.331.2:621.386.842

An X-Ray Image Amplifier using an Image Orthicon Camera Tube—E. Garthwaite and D. G. Haley. (J. Brit. IRE, vol. 19, pp. 615– 622; October, 1959. Discussion, pp. 622-623.) The special requirements of a television camera for use as an X-ray image amplifier are outlined, and details given of the special camera tube that has been developed.

621.398:551.507.362.2 629 Tracking Weather Satellites—(*Electronics*, vol. 32, p. 51; November 13, 1959.) A system for following and interrogating Tiros meteorological earth satellites is briefly described.

656.1:621.396.969.14 630 Measurement and Recording of Vehicle Speeds by means of Microwaves—H. Bürkle. (Arch. tech. Messen, no. 275, pp. R165–R168; December, 1958.) German battery-operated traffic radar equipment working at 3 cm λ is described.

PROPAGATION OF WAVES

621.371 631 Transfer of Transient Electromagnetic Surface Waves into a Lossy Medium—J. Keilson and R. V. Row. (J. Appl. Phys., vol. 30, pp. 1595–1598; October, 1959.) A flat earth with uniform electrical properties is assumed and the ionosphere disregarded in the treatment which gives a solution showing characteristics of wave propagation and diffusion. The limits imposed by the losses on the signal bandwidth are discussed.

621.391.81.029.4

On Propagation Velocity of Electromagnetic Waves at Audio Frequency—Ya, L. Al'pert and S. V. Borodina. (*Radiotekh. Elektron.*, vol. 4, pp. 195–201; February, 1959.) Determination of the phase velocity of em waves in the range 1–20 kc by means of harmonic analysis of atmospherics and their phase characteristics. The results of the investigation can be used to determine the effective conductivity of the lower ionosphere.

621.391.81.029.51./.53

Further Studies of the Deviation of Lowand Medium-Frequency Ground Waves at a Coast-Line—B. G. Pressey, G. E. Ashwell and R. Roberts. (*Proc. 1EE*, Part B, vol. 106, pp. 548–554; November, 1959.) Previous work is reviewed, and further experiments are described in which the transmitters were located at sea and directional measurements made on several sites on land. The deviation due solely to change in conductivity at the boundary is small compared with random errors attributed to ground irregularities and obstructions.

621.391.812.61.029.64

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Investigation of the Character of Rapidly Fading Radio Signals along a Transmission Path of Medium Length above the Earth's Surface-A. A. Semenov and G. A. Karpeev, (Radiotekh. Elektron., vol. 4, pp. 187-194; February, 1959.) A preliminary estimation of the effect of the underlying earth surface on the amplitude fluctuations of the reflected signal. The investigation is carried out by means of a 65-kw transmitter operating in the 3-cm band using a parabolic antenna with horizontal polarization. Signals are reflected from two standard reflectors located 15 and 36 km from the transmitter, one 30 meters higher than the other. Amplitude fluctuations are recorded on a cine film and graphs show the correlation with wind-velocity data during a period of 4 months.

621.391.812.62 635 Inadequacy of Scatter Mechanisms in Tropospheric Radio Propagation—P. C. M. de Belatini. (*Nature*, vol. 184, suppl. no. 20, pp. 1558–1559; November 14, 1959.) From a study of experimental results [*e.g.*, 2000*p* of 1959 (Kitchen, *et al.*)] it is concluded that the spatial field distribution is not random but essentially coherent. The fluctuations at a fixed point are due to shrinkage and expansion of the interference pattern with atmospheric changes.

621.391.812.621.029.64 636 Comparison of Computed with Observed Atmospheric Refraction—W. L. Anderson, N. J. Beyers and B. M. Fannin. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 258–260; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.391.812.622.029.64

Influence of an Atmospheric Duct on Microwave Fading—F. Ikegami. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 252–257; July, 1959, Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.391.812.623

Diffraction Theory of Tropospheric Propagation Near and Beyond the Radio Horizon: Parts 1 and 2--O. Tukizi. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 261-273; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.) See 3092 of 1959.

621.391.812.624:621.396.677

The Filling in of an Antenna Null by Off-Path Scattering on a Tropospheric Scatter Circuit—H. Staras. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 277-279; July, 1959. Abstract, PROC. IRE, vol. 47, pp. 2037-2038; November, 1959.)

621.391.812.624:621.396.96

Subhorizon Radar Echoes by Scatter Propagation—D. Atlas. (J. Geophys. Res., vol. 64, pp. 1205–1218; September, 1959.) An extensive combination of diffuse and striated echoes out to a maximum distance of 85 miles was observed on a 3-cm 300-kw radar in Kansas for about 4 hours in September, 1956. Neither direct back scatter from the atmosphere nor superrefraction can explain the features of the display. The phenomenon is attributed to ground back scatter. The height and characteristics of the scattering centers which would be necessary to uphold this explanation are unlikely, but it is pointed out that the phenomenon is rare; a gross estimate of the rate of occurrence is 1 day in 10⁴.

621.391.812.624.029.64

Microwave Scattering by Turbulent Air— C. E. Phillips. (IRE TRANS. ON ANTENNAS AND PROPAGATION, Vol. AP-7, pp. 245-251; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.391.812.63

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Rhythmic Fading of Short-Wave Radio Signals—B. N. Singh and R. L. Ram. (J. Atmos. Terr Phys., vol. 16, pp. 145–155; October, 1959.) The problem of fading is considered as equivalent to the superposition of two or more simple harmonic vibrations of similar frequencies and the results are applied to cases of interterence between magneto-ionic components. Typical fading curves are considered with particular attention to those cases in which the MUF for the F_2 layer passes through the signal frequency.

621.391.812.63

Long-Wave Field Enhancement and Short-Wave Fading—A. Haubert. (J. Atmos. Terr. Phys., vol. 13, pp. 379–381; February, 1959. In French.) The coincidence of fading on 6.2 ne with sudden field-strength fluctuations of 200-kc transmissions received in Rabat, Morocco, is investigated and an interpretation is given (see also 648 below). Recording of 200 kc transmissions may provide a better indication of SID than the present method of recording atmospherics at 27 kc.

621.391.812.63

Polarization Characteristics of Radio Wave Propagation in the Ionosphere—Y. S. N. Murty. (*Science and Culture (Calcutta*), vol. 25, pp. 161–162; August, 1959.) Expressions representing polarization characteristics obtained from Appleton-Hartree ray formulas and from the wave formulas of Saha, et al. (243 of 1952) are shown to be identical.

621.391.812.63 645

The Refraction of Radio Waves by a Spherical Ionized Layer—E. Woyk (Chvojková). (J. Atmos. Terr. Phys., vol. 16, pp. 124-135; October, 1959.) A general expression is derived for the refraction of a radio wave propagated obliquely in a spherically ionized layer. Both transmitted and reflected waves are considered. The special case of propagation along two fixed levels in an ionized layer appears naturally in the derivation of this expression.

621.391.812.63

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Transient Modes of High-Frequency Radio Wave Propagation across the Auroral Zone— B. J. Fulton, L. E. Petrie and W. S. P. Ward, (J. Atmos. Terr. Phys., vol. 16, pp. 185–186; October, 1959.) These modes are clearly defined and appear at the same time as the normal single-hop modes. They have higher MUF's and greater time delays than the normal modes and are probably due to irregularities in the ionosphere along the great-circle path.

621.391.812.63:539.16

H-Bomb Explosion Effects on Radio Communication—S. G. Kingan. (Short Wave Mag., vol. 17, pp. 321–322; October, 1959.) The influence of the two explosions in August, 1958, on ionospheric radio communications is discussed in relation to explosion height.

621.391.812.63:551.510.535:523.75 648 The Interpretation of Sudden Field Anomalies in the Long-Wave Range during Solar Flare Effects—E. A. Lauter and P. Třiska. (Z. Meteorol., vol. 13, pp. 190–192; July/August, 1959.) Field-strength measurements were made on 155 kc simultaneously at distances 1360 and 950 km from the transmitter to investigate the effect noted by Haubert (643 above). The anomalies appear to be due to phase changes between ground and sky wave where the virtual height of reflection drops.

621.391.812.63.029.45 649 VLF Reflections from the Ionosphere in the

VLF Reflections from the Ionosphere in the Presence of a Transverse Magnetic Field—

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MATHEMATICS

512:621.316.5 596 Algebraic Topological Methods for Contact Network Analysis and Synthesis-C. Saltzer. (Quart. Appl. Math., vol. 17, pp. 173-183; July, 1959)

512:621.318.57:681.142

Classification and Minimization of Switching Functions: Part 2-N. C. de Troye. (Philips Res. Repts., vol. 14, pp. 250-292; June, 1959.) Part 1-3048 of 1959.

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598 A Note on Addition Theorems for Mathieu Functions-K. Saermark. (Z. angew. Math. u Phys., vol. 10, pp. 426-428; July 25, 1959. In English.) Note on an addition theorem differing from that given by Meixner and Schäfke (511 of 1956).

MEASUREMENTS AND TEST GEAR

529.786 + 531.71 (083.7) 500 Atomic Standards of Length and Time-H. Barrell and L. Essen. (Sci. Progr., vol. 47, pp. 209-229; April, 1959.) The development of sources of monochromatic light and atomic beams is reviewed, and the design and application of spectral-line frequency standards are described. 31 references.

621.3.018.41(083.74)

Using Low-Frequency Standard Broadcasts -H. F. Burgess and M. C. Jones. (Electronics, vol. 32, pp. 48-49; October 30, 1959) A technique is outlined for the calibration of 100-kc oscillators, from signals of the NBS 60-kc standard-frequency transmission (Station KK2XEI). The output of a coherent detector gives a continuous record of error, and the use of a narrow bandwidth permits operation at low input signal levels. Errors can be reduced to the order of 5 parts in 1010.

621.317.3:621.382.2

The Measurement of Semiconductor Diode Switching Characteristics-J. N. Barry and S. F. Fisher. (Brit. Commun. Electronics, vol. 6, pp. 788-791; November, 1959.) Equipment for measuring the transient reverse current which flows when the applied voltage is switched from forward to reverse.

621.317.331.087.6:537.311.33 602 Improved Automatic Four-Point Resistivity Probe-D. Dew-Hughes, A. H. Jones and G. E. Brock. (Rev. Sci. Instr., vol. 30, pp. 920-922; October, 1959.) Apparatus is described which automatically measures and plots the resistivity at fixed intervals along a semiconductor sample. The probe point spacing is 0.025 inch.

621.317.335+621.317.41.029.64:621.318.134 603

Apparatus for the Measurement of Tensor Permeability and Dielectric Properties of Ferrites at X-Band Frequencies-W. S. Carter. (Marconi Rev., vol. 22, pp. 154-163; 3rd Quarter, 1959.) Equipment is described for measuring dielectric constant and loss on a routine basis. Measurements of permeability are also possible on representative samples.

621.317.337:621.372.413 604 The Measurement of the O-Factor of Cavity Resonators Coupled to Transmission Lines, using a Measurement Line-H. W. Urbarz. (Nachrichtentech. Z., vol. 11, pp. 571-576; November, 1958.) Expansion of the method of measurement given in 3141 of 1956.

621.317.4

Magnetic Measurements-K. J. Chou-dhury and P. C. Sen. (Electronic Radio Eng., vol. 36, pp. 422-426; November, 1959.) A description is given of a modified bridged-T network for the measurement of incremental magnetic loss and ac permeability of cores subjected to superposed direct and alternating flux. Theoretical equations are presented for use over the frequency range 50 cps-50 kc.

621.317.4:538.632

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Multiple-Element Hall-Effect Sensor-M. Epstein, H. M. Sachs and J. L. Greenstein. (PROC. IRE, vol. 47, p. 2014; November, 1959.) A design suiatble for magnetic-field measurements

621.317.444:550.380.8

Some Remarks on the Proton Magne-tometer—G. Klose. (Z. angew. Phys., vol. 10, pp. 495-497; November, 1958.) The principle of operation of the magnetometer [see, e.g., 526 of 1957 (Cahill and Van Allen)] and the accuracy of the measurement of the earth's magnetic field are discussed.

621.317.74:621.372.8

Recording Attenuation of Waveguide Components-G. Edelcreek. (Electronics, vol. 32, p. 126; October 23, 1959.) Description of a frequency-sweep method for measuring attenuation up to 80 db, with photographic recording of the CRO trace.

621.317.742 609 Unconventional Technique for Measuring VSWR-J. Hanson. (Electronics, vol. 32, pp. 120-121: October 23, 1959) A transistorized instrument comprising an oscillator, a RC directional coupler and a meter indicator is described for measurements in the range 150-175 mc.

621.317.75 610 An Amplitude Distribution Meter-M. Drayson. (*Electronic Engrg.*, vol. 31, pp. 578– 584; October, 1959.) The significance of signal amplitude distribution measurements is described, with a brief outline of the mathematical expressions and approximations involved in any practical system. The system described uses a specially developed CRT as the waveform sampler and has a resolution of 1 per cent in the band 0.1 cps-10 mc over a 0.3-30-volt amplitude range.

621.317.75:621.395.625.3

The Influence of Recording-Head and Tape Properties on the Recording of Magnetic-Tape Oscillograms-W. Reinert. (Elektronik, vol. 7, pp. 329-335; November, 1958.) Design problems of the magnetic-tape oscillograph (2187 of 1958) are d scussed. A four-track recording head is described with illustrations of oscillograms produced by it and by a light-beam oscillograph for comparison.

621.317.755.087.6

New Method for Graphical Reproduction of Cathode-Ray Oscillograms-R. K. Swank and E. A. Mroz. (Rev. Sci. Instr., vol. 30, pp. 880-884: October, 1959.) Description of an automatic optical-electronic device using two photomultipliers in a null system to reproduce graphically to high accuracy a repetitive CRO trace.

621.317.772.029.64:621.396.65 613 Experimental Equipment for Measuring Group Delay in the Frequency Band 3.8-4.2 Gc/s-R. J. Turner. (P.O. Elec. Engrg. J., vol. 52, Part 53, pp. 207-211; October, 1959.) Measures group delay to within ± 0.2 mµsec in 50 musec by determining the phase shift of a 1-mc modulating signal (low-deviation FM). A balanced phase comparator using thermionic diodes with adjustable heater voltages permits phase determination at 1 mc to within $\pm 0.07^{\circ}$.

621.317.79:681.142:621.385.833

An Electon-Trajectory Tracer for use with the Resistance Network Analogue-M. E. Haine. (Proc. IRE, Part B, vol. 106, pp. 517-525; November, 1959. Discussion, pp. 525527.) "The paper describes an instrument for direct analog computation of electron trajectories, a resistance network providing the necessary field data. Constructional details are given and results for two typical electrostatic lenses are shown and compared with results obtained experimentally. Methods for improving accuracy and speed of operation are outlined.

621.317.794

The Inherent Sensivitity of Metal Bolometers-G. Barth. (Optik, vol. 15, pp. 694-709; November, 1958.) The maximum sensitivity of a bolometer is calculated in a bridge circuit designed for optimum conditions, taking account of the limit set by thermal noise and temperature fluctuations in the ambient around the bolometer foil. Optimum design features and operating conditions are summarized.

021.317.799:629.19

Electronic Instruments in Space-Research Vehicles-R. L. F. Boyd, (J. IEE, vol. 5, pp. 457-463; August, 1959.) The development of rocket-borne instrumentation is discussed with reference to the problems of special environmental conditions, power supply, and weight and size restrictions. The payloads of seven space vehicles are compared.

OTHER APPLICATIONS OF RADIO AND ELECTRONICS

531.74:621-52:621.396.934 617 High-Resolution Angle Transducer and Encoder-L. G. de Bey, D. Comstock, S. B. Peterson and R. C. Webb. (Electronics, vol. 32, pp. 78-81; October 16, 1959.) Equipment is described which will measure shaft position in missile tracking instruments to an accuracy within 0.001° for rotation rates up to 100°/sec-

621.362:537.58 618 Thermoelectron Engines: Future Power Sources?—G. N. Hatsopoulos, J. Welsh and E. Langberg. (Electronics, vol. 32, pp. 69-72; November 13, 1959.) The basic principles and

possible heat sources are reviewed. Three meth-

ods of reducing the space-charge potential

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Microwave Generators with Enclosed Work Chambers for the Dielectric Heating of Food and Industrial Products-W. Schmidt. (Elek tron. Rundschau, vol. 12, pp. 390-393 and 417-420; November and December, 1958; and vol. 13, pp. 13-16; January, 1959.) The theory of a magnetron HF generator, its design and practical construction, and related measurement problems are discussed.

621.38:[57+61

barrier are discussed.

621.365.55:621.373.421.14.029.6

620 Biomedical Electronics-(Proc. IEE, vol.

47, pp. 1815-2010; November, 1959.) A collection of 25 papers relating to electronic applications in biology and medicine.

621.384.62

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New Electrostatic Accelerator-I. Michael, E. D. Berners, F. J. Eppling, D. J. Knecht, L. C. Northcliffe and R. G. Herb. (Rev. Sci. Instr., vol. 30, pp. 855-863; October, 1959.) The design and construction of a short, bakeable accelerator is described.

621.385.832:681.142

Character Displays using Analogue Techniques-S. C. Chao. (Electronics, vol. 32, pp. 116-118; October 23, 1959.) Fast read-out is obtained by forming characters on a CRT display from a series of overlapping dots. Resistor summing networks establish the dot positions.

621.385.833 623 Comparative Considerations on the Aper-

ture Error of Symmetric and Asymmetric Uni-

1958) and give recombination data in agreement with those obtained by other methods.

537.311.33:546.289 Dislocation Pinning in n-Type Germanium -R. L. Cummerow and A. R. Cherry. (Phys. Rev. Lett., vol. 3, pp. 367-368; October 15, 1959.) A thermally-induced-glide technique was used on samples of p-type Ge doped with Ga and n-type Ge doped with As. A glide was observed on the p-type but none was present on the n-type until a temperature of 700°C was reached. An explanation of this pinning by the As impurities is given.

537.311.33:546.289:535.215-15 567 Emission and Absorption of Long-Wave Infrared Radiation by Germanium in the Photoconducting State—F. R. Kessler. (Z. Naturforsch., vol. 13a, pp. 295-302; April, 1958.) The absorption spectrum of electron-hole pairs in the range 3-15 microns was determined in Ge at temperatures between 20 and 70°C. A maximum of infrared emission was found at a wavelength of 10 microns; this was investigated for various temperatures and as a function of the number of electron-hole pairs. A phenomenological interpretation of the effects is given.

537.311.33:546.289:537.312.9568 Piezoresistance of *n*-Type Germanium-II. Fritzsche. (Phys. Rev., vol. 115, pp. 336-345; July 15, 1959.) A study at low temperatures of the departure of the piezoresistance from linear dependence on applied stress and a test of the predictions of the electron transfer model at large stresses. Results indicate that the model is a useful one.

537.311.33:546.289:548.73 569 X-Ray Representation of the Dislocation Field of Individual Dislocations in Germanium Single Crystals—U. Bonse and E. Kappler. (Z. Naturforsch., vol. 13a, pp. 348-349; April, 1958; plate.)

537.311.33:546.3-87'86 570 Semiconducting Properties of Bi-Sb Alloys -S. Tanuma. (J. Phys. Soc. Japan, vol. 14, p. 1246; September, 1959.) Measurements of resistivity, Hall effect and magnetoresistance of polycrystalline specimens, in the temperature range 4.2°K to 300°K.

537.311.33:546.36'59 571 Studies of the Semiconducting Properties of the Compound CsAu-W. E. Spicer, A. H. Sommer and J. G. White. (Phys. Rev., vol. 115, pp. 57-62; July 1, 1959.) Experimental results are given and discussed.

537.311.33:546.47-31:535.34-15 572 Infrared Absorption in Zinc Oxide Crystals D. G. Thomas. (J. Phys. Chem. Solids, vol. 10, pp. 47-51; April, 1959.) Absorption in the 1- to 12-micron band is shown to be due to free carriers, lattice vibration bands and photoionization of impurities.

537.311.33: 546.48'19 573 Cd₃As₂—a Noncubic Semiconductor with Unusually High Electron Mobility—A. J. Rosenberg and T. C. Harman. (J. Appl. Phys., vol. 30, pp. 1621-1622; October, 1959.) At a carrier concentration of 4×10^{18} /cm³ the mobility at 300°K is 10,000, the highest reported for any material.

537.311.33:546.561-31

The Semiconductor Properties of Cu2O-K. Stecker. (Ann. Phys. (Lpz.), vol. 3, pp. 55-81; February 28, 1959.)

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Part 12: The Conductivity of Cu₂O within the Existence Limits at High Temperatures in the Range of Low Pressures (pp. 55-69).

Part 13: Conductivity Measurements on

Cu₂O within the Existence Limits with a Disturbance of the Thermodynamic Equilibrium (pp. 70-81).

Part 11: 3595 of 1954 (Blankenburg).

575 537.311.33:546.68'19 Thermal Electrical and Optical Properties of (In,Ga)As Alloys—M. S. Abrahams, R. Braunstein and F. D. Rosi. (J. Phys. Chem. Solids, vol. 10, pp. 204-210; July, 1959.)

537.311.33:546.628'86 576 Recombination Processes in p-Type Indium Antimonide-R. N. Zitter, A. J. Strauss and A. E. Attard. (Phys. Rev., vol. 115, pp. 266-273; July 15, 1959.) Photoelectromagnetic and photoconductive lifetimes have been measured from 77° to 300°K in monocrystalline p-type InSb of net acceptor concentration ranging from less than 10¹⁵ cm⁻³ to 10¹⁵cm⁻³. Experimental procedures and results are discussed.

537.311.33:546.682'86 577 Electron Damage Thresholds in InSb-F. H. Eisen and P. W. Bickel, (Phys. Rev., vol. 115, pp. 345-346; July 15, 1959.) Measurements indicate that displacements are produced at electron energies as low as 240 kev.

537.311.33:546.873'241 578 Bismuth Telluride and Related Compounds -D. A. Wright, [Research (London), vol. 12, pp. 300-306; August/September, 1959.] The structure and physical properties are reviewed and reference is made to thermoelectric applications.

537.311.33:548.73

Shadows of Dislocation Lines in X-Ray Diagrams-G. Borrmann, W. Hartwig and H. Irmler. (Z. Naturforsch., vol. 13a, pp. 423-425; May, 1958.) Diagrams obtained with a Si disk are shown and discussed.

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580 537.32 Materials for Thermoelectric Refrigeration F. D. Rosi, B. Abeles and R. V. Jensen. (J. Phys. Chem. Solids, vol. 10, pp. 191-200; July, 1959.) Thermoelectric propreties of Bi2Te3 and alloys with Sb and Se were measured.

537.32 581 Theory of Thermoelectric Power of Ionic Crystals: Part 3-E. Haga. (J. Phys. Soc. Japan, vol. 14, pp. 1176-1181; September, 1959.) The variation with time is calculated of the thermoelectric power in a AgCl crystal doped with CuCl for prescribed conditions of temperature gradient in the crystal.

537.583 Characteristics of UC,ZrC and (ZrC)(UC) as Thermionic Emitters-R. W. Pidd, G. M. Grover, D. J. Roehling, E. W. Salmi, J. D. Farr, N. H. Krikorian and W. G. Witteman. (J. Appl. Phys., vol. 30, pp. 1575-1578; October, 1959.) Excellent emission properties are reported.

538:061.3 583 Magnetism and Magnetic Materials—(J.Appl. Phys., vol. 30, suppl., pp. 1S-323S; April, 1959.) The text is given of papers presented at a conference held in Philadelphia, Pa., November 17-20, 1958.

538.22:538.569.4 584 Structure-Sensitivity of the High-Frequency NMR in Powdered Antiferromagnetic MnF2-J. L. Davis, G. E. Devlin, V. Jaccarino and A. L. Schawlow. (J. Phys. Chem. Solids, vol. 10, pp. 106-109; July, 1959.)

538.22:538.569.4 585 Electron Spin Resonance of Gd3+ in Lanthanum Fluoride—D. A. Jones, J. M. Baker and D. F. D. Pope. (Proc. Phys. Soc., vol. 74, pp.

249-256; September, 1959.) The large crystalfield splitting observed in a single-crystal of LaF₃ containing 0.01 per cent Gd³⁺, which is about 0.3 cm⁻¹ in zero magnetic field, indicates a possible application of the salt as a maser material.

538.221:539.23 586 Curie Point in Thin Ni Films determined by Electrical Method-K. Kuwahara. (J. Phys. Soc. Japan, vol. 14, p. 1246; September, 1959.) Determinations of Curie temperature, using the anomalies in resistance and magnetoresistance which occur at that point.

538.221:621.318.124

Magnetic Materials with Perminvar Effect: Part 4-Perminvar and Magnetic-Field Annealing Effect in connection with the After-Effect in Ferrites due to Electron Diffusion-A. v. Kienlin, (Z. angew. Phys., vol. 10, pp. 562-565; December, 1958.) Part 3: 2324 of 1959.

588 538.221:621.318.134 Low-Temperature Heat Capacities and Thermodynamic Properties of Zinc Ferrites: Part 3-E. F. Westrum, Jr. and D. M. Grimes. (J. Phys. Chem. Solids, vol. 10, pp. 120-125; July, 1959.)

Part 1-3178 of 1958.

Part 2-J. Phys. Chem. Solids, vol. 6, pp. 280-286; August, 1958.

538.221:621.318.134 589 The Ferrimagnetism and Crystal Chemistry of Substituted Manganese-Tin Spinels-M. A, Gilleo and D. W. Mitchell. (J. Phys. Chem. Solids, vol. 10, pp. 182-186; July, 1959.)

538.221:621.318.134 590 The Interaction of Magnetic Ions in Gd₃Mn₂Ge₂GaO₁₂ and Related Garnets-M. A. Gilleo and S. Geller. (J. Phys. Chem. Solids, vol. 10, pp. 187-190; July, 1959.)

538.221:621.318.134 501 Proposed Means for Realizing High Power Stability in Magnetic Oxides-L. G. Van Uitert, R. C. LeCraw, E. G. Spencer and R. L. Martin. (J. Appl. Phys., vol. 30, pp. 1623– 1624; October, 1959.) The stability is proportioned to the line width (ΔHk) of the spin wave that first goes unstable as the RF power is increased; possible means of increasing ΔHk are suggested.

538.222:538.569.4 592 Quadrupole Selection Rule in Iron-Group Spin-Phonon Interactions-R. D. Mattuck and M. W. P. Strandberg, (Phys. Rev. Lett., vol. 3, pp. 369-370; October 15, 1959.)

539.2:539.12.04

ences.

593 Mechanical Properties of Irradiated Solids -F. A. Levi. (Nuovo Cim., vol. 12, suppl. no. 2, pp. 123-295; 1959.) Review of literature dealing with radiation effects in solids. 693 refer-

621.315.6:537.311 594 Two-Carrier Space-Charge-Limited Current in a Trap-Free Insulator-R. H. Parmenter and W. Ruppel. (J. Appl. Phys., vol. 30, pp. 1548-1558; October, 1959.)

621.318.132:621.375.3.042.143 595 Core Materials and Core Designs for Magnetic Amplifiers-H. Fachse. (VDI Zeitschrift, vol. 101, pp. 341-342; March 21, 1959.) Review of modern lamination materials, particularly of those with grain orientation, including an outline of design features required for full exploitation of magnetic properties.

ous kinds of crystal are formed.

535.376

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540 The Phasing of Luminescence-Wave Secondary Maxima of Electroluminescence-D. Hahn and F. W. Seemann. (Z. Naturforsch., vol. 13a, pp. 349-350; April, 1958.) The tempera-

ture dependence of the phase of secondary waves with sinusoidal excitation is investigated. The apparently large phase shift may be due to an incorrect interpretation of measurements. See also 833 of 1959.

535.376

Long-Period Afterglow of KCI: Tl Phosphor under Cathode-Ray Excitation-V. V. Ratnam. (Proc. Nat. Inst. Sci. India, Part A. vol. 25, pp. 111-117; March 26, 1959.) KCI: TI phosphor was excited by 11-ky cathode rays and the decay of the afterglow studied over the temperature range of 150-400°K. Results indicate that the value of s, the frequency factor, is not the same for all trapping centers.

535.376:546.47'221

Electroluminescence of ZnS Phosphors Excited by Short Field Pulses-S. Tanaka. (J. Phys. Soc. Japan, vol. 14, pp. 1123-1140; September, 1959.) Experimental studies of emission spectra, brightness waveforms, decay times, and dependence of light output on applied voltages are described. The results are discussed in relation to the mechanisms involved and the corresponding effects with sinusoidal field excitation.

537.226 543 Dielectric Research-(Tech. News Bull. Nat. Bur. Stand., vol. 43, pp. 168-174; September, 1959.) A report of studies of a wide range of dielectric materials.

537.227

Built-In Nucleation Sites in Triglycine Sulphate-A. G. Chynoweth and J. L. Abel. (J. Appl. Phys., vol. 30, pp. 1615-1617; October, 1959.) Sites for domain nucleation at low fields are mainly determined by gross singularities or conditions built in as the crystal grows.

537.227:546.431'824-31 545 Ferroelectric After-Effects in Polycrystalline Barium Titanate-W. Koch. (Z. Naturforsch., vol. 13a, pp. 303-310; April, 1958.) The anomalous variation of dielectric constant when an electric field is applied to BaTiO₃ specimens or when the specimens are subjected to hydrostatic pressure is investigated. Results are interpreted in terms of the splitting up of domains and the shifting of the tetragonal-to-orthorhombic transition temperature.

537.227:546.431'824-31:537.311.33 546 Properties of Semiconductive Barium Titanates-O. Saburi. (J. Phys. Soc. Japan, vol. 14, pp. 1159-1174; September, 1959.) The reduction of the resistivity of barium titanate by various additives is investigated, and mechanisms for the effect are discussed.

537.227:546.431'824-31:538.569.4

Electron Paramagnetic Resonance in Single Crystals of BaTiO3-A. W. Hornig, R. C. Rempel and H. E. Weaver. (J. Phys. Chem. Solids, vol. 10, pp. 1-11; April, 1959.) The resonance spectrum is due to the impurity ion Fe3+, believed to be located at the titanium position in the unit cell.

537.227:546.431'824-31:539.12.04 548 Radiation-Induced Changes in the Ferroelectric Properties of some Barium-Titanate-

Type Materials-I. Lefkowitz. (J. Phys. Chem. Solids, vol. 10, pp. 169-173; July, 1959.) Measurements on pure ceramic BaTiO3 showed a depression of the dielectric constant peak with little modification of the room temperature dielectric constant. Ceramics made with additives showed a shift of the Curie point as well as a depression of the dielectric constant peak. Materials exposed to an integrated pile dosage of 1×10¹⁸nvt would not support reversible dielectric polarization.

537.311.31:539.23:537.228

Surface States in Metals-G. Bonfigioli and R. Malvano. (Phys. Rev., vol. 115, pp. 330-335; July 15, 1959.) Conductivity modulation by an electric field has been measured in Au, Sb and Bi films using a new technique. The existence of localized and conducting "surface states" at the metal/dielectric interface seems to be experimentally confirmed.

537.311.31:621.317.321

The Contact Potential on Metal Surfaces with Oxidation and with Adsorption-W. Schaaffs. (Z. angew. Phys., vol. 10, pp. 503-511; November, 1958.) For the method of measurement used, see 225 of January (Schaaffs and Woelk).

537.311.33

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Organic Semiconductors-D. D. Elev. [Research (London), vol. 12, pp. 293-299; August/September, 1959.] Results of measurements of electrical conductivity of crystalline organic substances are reviewed. Conductivity may be associated with the intermolecular tunneling of thermally excited π -electrons, 54 references.

537.311.33

552 Theory and Application of a Minority Carrier Sweep-Out Effect-R. D. Larrabee. (J. Appl. Phys., vol. 30, pp. 1535-1538; October, 1959.) A simplified analysis shows how the effect may be used to determine the density and drift mobility of carriers of both signs.

537.311.33

Determination of Avalanche Breakdown in p-n Junctions-J. Maserjian. (J. Appl. Phys., vol. 30, pp. 1613-1614; October, 1959.) A single effective value of the ionization rate per cm for both holes and electrons may be used in the approximate analysis presented. See also 2453 of 1958 (Chynoweth).

537.311.33:537.32

On the Theory of Thermoelectricity-J. Tauc. (J. Phys. Soc. Japan, vol. 14, pp. 1174-1175; September, 1959.) Discussion of the electric field inside a two-band semiconductor with temperature-dependent energy gap. See 3015 of 1959 (Haga).

537.311.33:537.32:538.63 555 Theory of Thermomagnetic Effects of Nonpolar Isotropic Semiconductors-J. Appel. (Z. Naturforsch., vol. 13a, pp. 386-402; May, 1958.) Changes of transverse thermoelectric power, thermal conductivity and Nernst-Ettingshausen coefficient are calculated as a function of temperature and magnetic field strength. Good agreement with theory can be obtained for the results of measurements on pure p-type Ge at 80°K. See also 3511 of 1957 (Parrott).

537,311.33:546.26-1

Hall Coefficient and Magnetoresistance in Semiconducting Diamond-R. T. Bate and R. K. Willardson. (Proc. Phys. Soc., vol. 74, pp. 363-367; September 1, 1959.) The Hall coefficient was found to increase monotonically with increasing magnetic field H, while the transverse magnetoresistance was proportional to H^2 at low-field strengths.

537.311.33:546.28

A Volume Effect in the Etching of Silicon Single Crystals-H. Benda. (Z. Naturforsch., vol. 13a, pp. 354-355; April, 1958.) The effect observed may be due to the diffusion of hydrogen into the silicon.

537.311.33:546.28 558

Birefringence due to Residual Stress in Silicon-J. Hornstra and P. Penning. (Philips Res. Repts., vol. 14, pp. 237-249; June, 1959.)

537.311.33:546.28

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Cleaning of Silicon Surfaces by Heating in High Vacuum-F. G. Allen, J. Eisinger, H. D. Hagstrum and J. T. Law. (J. Appl. Phys., vol. 30, pp. 1563-1571; October, 1959.) Heating at 1550°K for several minutes forms permanent p-type layers several microns deep with atomically clean surfaces.

537.311.33:546.28

560 Influence of the Ambient Atmosphere on the Surface Recombination of Silicon-H. U. Harten. (Philips Res. Repts., vol. 14, pp. 207-210; June, 1959.) Measurements of the surface photovoltage show that the surface potential of Si can be altered over a wide range by chemical surface treatments and over a smaller range by the ambient atmosphere. An investigation of the surface recombination shows this process to be determined chiefly by recombination centers of the "Hall-Shockley-Read-type.

537.311.33:546.289

Pressure-Dependence of the Resistivity of Germanium-A. Michels, J. Van Eck, S. Machlup and C. A. Ten Seldam. (J. Phys. Solids, vol. 10, pp. 12-18; April, 1959.) "The effect of hydrostatic pressure on the resistance of a p-type sample of germanium (specific resistivity about 80 Ω-cm at 293°K) has been investigated up to 2700 atm. at temperatures between 125 and -150°C. The results in the intrinsic range indicate an increase in the energy gap of 5.4×10⁻⁶ ev/atm, in agreement with earlier experimental determinations. In the extrinsic region, the resistivity decreases slightly with pressure, indicating an increase in hole mobility of 9 ppm/atm.

537.311.33:546.289 562

The Thermal Conductivity of Germanium at High Temperatures-F. Kettel. (J. Phys. Chem. Solids, vol. 10, pp. 52-58; April, 1959. In German.)

537.311.33:546.289 563 Recombination Properties of Nickel in Germanium-G. K. Wertheim. (Phys. Rev., vol. 115, pp. 33-47; July 1, 1959.) Experimental lifetime data agree with known energy levels and solid solubility. Three electron-capture cross sections associated with the three charge states assigned to Ni have been determined.

537.311.33:546.289

Thermal Oscillations in n-Germanium at Low Temperature—S. H. Koenig and R. D. Brown, III. (J. Phys. Chem. Solids, vol. 10, pp. 201-203; July, 1959.) Instability occurs when a small increase in current in part of the sample produces a local temperature rise which cannot be dissipated before it in turn causes a further current increase. The oscillations cause an effective negative resistance to appear when dc measurements are made in the usual manner.

537.311.33:546.289

Recombination Relaxation Effects in Germanium Surfaces-D. H. Lindley and P. C. Banbury. (Proc. Phys. Soc., vol. 74, pp. 395-400; October 1, 1959.) A small electric field of variable frequency was used to modulate the conductance of a thin crystal slab. The resulting dispersion has been studied and the relauts are in agreement with Garrett's model (153 of

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1958 were used to derive the mean-electrondensity profile. The electron density falls from about 0.7×10^{12} meter⁻³ at 380 km to half this value at about 660 km, and to 0.15×10^{12} meter⁻³ at a height of 1200 km.

551.510.535:621.391.812.63 513 Sudden Changes in the Virtual Height of Radio Waves Reflected from the *E* Region of the Ionosphere—J. D. Whitehead. (*J. Atmos. Terr. Phys.*, vol. 16, pp. 99–102; October, 1959.) It is shown that the lower-level reflections are from thin kyers within the normal *E* region. The changes in amplitude are used to deduce the thickness of these layers and the electronic collision frequency in the *E* region.

551.510.535:621.391.812.63:523.75 514 The Interpretation of Sudden Field Anomalies in the Long-Wave Range during Solar Flare Effects—Lauter and Třiska. (See 648.)

551.510.535(98) The Height of F-Layer Irregularities in the Arctic Ionosphere—II. F. Bates. (J. Geophys. Res., vol. 64, pp. 1257-1265; September, 1959.) Frequency-sweep back scatter soundings from College, Alaska, show that F-layer irregularities exist at heights of 350-500 km, and sometimes extend 1500 km in a north-south direction.

551.510.535(99)

Observations of the Ionosphere over the South Geographic Pole—R. W. Knecht. (J. Geophys. Res., vol. 64, pp. 1243–1250; September, 1959.) "It is found that F-region ionization persists throughout the 6-month winter night. Marked diurnal variations are observed in the monthly medians of f_0F_2 even though the usual daily variation in solar elevation is absent at this unique location. A small but significant diurnal variation is also found in f_0F_1 . In contrast, f_0F_2 exhibits no regular daily fluctuation, but seems to depend to a greater extent on the level of solar activity."

551.510.536:550.38

Motions in the Magnetosphere of the Earth -T. Gold. (J. Geophys. Res., vol. 64, pp. 1219–1224; September, 1959.) The magnetosphere is defined as the region above the ionosphere in which the earth's magnetic field has dominant control over the motions of gas and fast charged particles. Conditions in the magnetosphere and the dynamical behavior of its constituents are discussed.

551.594.5

Horizontal Movements of Visual Auroral Features—S. Evans. (J. Atmos. Terr. Phys., vol. 16, pp. 191–193; October, 1959.) Measurements with all-sky cameras show that, for Halley Bay during July and August, 1956, auroral movements were predominantly eastwest with a reversal in direction from west to east at 0300 UT.

551.594.5

Existence of an Inner Auroral Zone—K. Lassen. (*Nature*, vol. 184, suppl. no. 18, pp. 1375–1377; October 31, 1959.) Study of a "population" of auroras which seems to form an inner auroral zone.

551.594.5

Type-B Aurora in the Antarctic—J. M. Malville. (J. Atmos. Terr. Phys., vol. 16, pp. 59-66; October, 1959.) The results of auroral observations are discussed and an excitation mechanism is suggested to explain the type-B spectrum.

551.594.5:621.396.96 VHF and UHF Radar Observations of the Aurora at College, Alaska—R. I. Presnell, R. L. Leadabrand, A. M. Peterson, R. B. Dyce, J. C. Schlobohm and M. R. Berg. (J. Geophys. Res., vol. 64, pp. 1179–1190; September, 1959.) Within the frequency interval 200–800 mc, radar echoes show that the aurora conforms to the Booker model (27.39 of 1956), provided that the longitudinal correlation length is halved and the transverse correlation distance is reduced from 0.16 meter to 0.1 meter. A daytime "diffuse" aurora is found to exist in the E layer, almost parallel to the earth's surface, and is under strong solar control.

551.594.5:621.396.96

High-Altitude 106.1-mcs Radio Echoes from Auroral Ionization Detected at a Geomagnetic Latitude of 43°—J. C. Schlobohm, R. L. Leadabrand, R. B. Dyce, L. T. Dolphin and M. R. Berg. (J. Geophys. Res., vol. 64, pp. 1191-1196; September, 1959.)

551.594.5:621.396.96

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Doppler Investigations of the Radar Aurora at 400 Mc/s—R, L. Leadabrand, R. I. Presnell, M. R. Berg and R. B. Dyce. (J. Geophys. Res., vol. 64, pp. 1197–1203; September, 1959.)

551.594.6 524 Directional Observations of Radio Noise from the Outer Atmosphere—G. R. A. Ellis and D. G. Cartwright. (*Nature*, vol. 184, suppl. no. 17, pp. 1307–1308; October 24, 1959.) Report of preliminary results obtained with a direction finder operated at a wave-frequency of 4.5 kc.

551.594.6 525 Spaced Observations of Radio Noise from the Outer Atmosphere—G. R. A. Ellis, D. G. Cartwright and J. R. V. Groves. (*Nature*, vol. 184, suppl. no. 18, pp. 1391–1392; October 31, 1959.) Observations made at two stations 1000 km apart show that the regions in which the noise is generated are normally stationary with respect to the earth.

551.594.6 526
Location of Initial Sferics of Long Whistlers
—G. Entzian and C. Popp. (Z. Meteorol., vol. 13, pp. 193-194; July/August, 1959.) Sources of atmospherics were located by a U.S.S.R. research vessel in the vicinity of Ireland in December, 1958, and coincidences with long whistlers were observed. One of the records is analyzed.

551.594.6:551.510.535

Synthesis of the Waveforms of Atmospherics and Effective Parameters of the Lower Ionosphere at Low Frequencies—Ya. L. Al'pert and D. S. Fligel'. (*Radiotekh. Elektron.*, vol. 4, pp. 202–211; February, 1959.) Theoretical estimations of the waveforms of atmospherics are compared with signals received over a distance of 500–3000 km. Results are tabulated for the frequency range 5–10⁵ cps.

LOCATION AND AIDS TO NAVIGATION 621.396.9:656.052:061.3 528

Automatic Methods of Navigation—(J. Inst. Nav., vol. 12, pp. 318–333; July–October, 1959). A list is given, with summaries, of papers presented at the Convention held in Paris, April 28-29, 1959.

621.396.932.2:523.164.3

Automatic Radio-Celest al Navigation— G. R. Marner. (J. Inst. Nav., vol. 12, pp. 249– 259; July-October, 1959.) A general discussion of the problems associated with radiocelestial navigation and of the optimum frequency for a practical system using the sun and moon as radiation sources.

621.396.933.1

Air and Sea Tests of the Dectra Radio-Navigation System—C. Powell. (J. Inst. Nav., vol. 12, pp. 289-307; July-October, 1959.) A summary of the results obtained during the first two years' operation of the experimental Dectra chain in the North Atlantic area. Reference is made to observations at fixed monitor stations and to the data link for airto-ground transmission of the Dectra fix.

621.396.96:621.391.812.624 531 Subhorizon Radar Echoes by Scatter Propagation—Atlas. (See 640.)

MATERIALS AND SUBSIDIARY TECHNIQUES

535.215:535.37

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Calculation of the Photoconductivity from A.C. Impedance Changes Induced in ZnS and ZnCdS Phosphors—H. Kallmann, B. Kramer and P. Mark. (J. Phys. Chem. Solids, vol. 10, pp. 59–63; April, 1959.) Equations are derived from which the photoconductivity and conduction-band electron density can be obtained from the ac measurements. The theoretical and experimental results for five phosphors are compared.

535.215:537.311.33 533 Optical Sensitization in the Photoelectric

Effect at the Contact between a Semiconductor and an Organic Dye—E. K. Putseiko, (Dokl. Akad. Nauk SSSR, vol. 129, pp. 303–306; November 11, 1959.) This effect can be obtained by absorption of the dye from a solution or by pressing the solid layers of the dye powder against the ZnO semiconductor. Results indicate that the maximum photo-emf at the junction of dye and ZnO is of several millivolts per milliwatt.

535.215:546.48'221

Edge Photoconductivity of Cadmium Sulphide—M. Avinor. (*Philips Res. Rept.*, vol. 14, pp. 211–214; June, 1959.) It is shown that the characteristic photoconductivity peak of single crystals of CdS at 515 m μ is not due to band-band transition.

535.215:546.48'221

Induced Conductivity of CdS by β - and γ rays—S. Ibuki. (*J. Phys. Soc. Japan*, vol. 14, pp. 1196–1204; September, 1959.) Experimental study, and comparison with effects produced by visible light.

535.215:546.492'151 536 The Influence of the Contacts on the Photoconductivity of Red Mercury Iodide—E. Batt and F. Stöckmann. (*Z. Naturforsch.*, vol. 13a, pp. 352–354; April, 1958.) Discrepancies between the I/V characteristics reported by different authors (e.g., R. II. Bube, *Phys. Rev.*, vol. 106, pp. 703–717; May 15, 1957) are shown, by measurements, to be due to the effect of different contact materials.

535.37 537 A Theory of Edge-Emission Phenomena in CdS, **ZnS and ZnO**—J. J. Hopfield. (*J. Phys. Chem. Solids*, vol. 10, pp. 110–119; July, 1959.) Based on a tight-binding model, hand symmetries and splittings which reproduce the currently known polarization phenomena in absorption and edge emission are given. Experimental observations are compared with predictions based on this model.

535.37:546.48'221 538

Edge and Impurity Emission in Cadmium Sulphide—D. Warschauer and D. C. Reynolds. (*Phys. Rev. Lett.*, vol. 3, pp. 370-372; October 15, 1959.) Some observations are given which cannot be explained in terms of the simple model recently proposed [*e.g.*, 2268 of 1959 (Birman)].

535.37:546.48'221:548.5 539 On the Crystal Growth of Cadmium Sul-

phide-S. Ibuki. (J. Phys. Soc. Japan, vol. 14,

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World Radio History

232; November, 1958.) Four types of radiosonde are described.

551.51+523.755 491 The Earth in the Sun's Atmosphere-S. Chapman. (Sci. Amer., vol. 201, pp. 64-71; October, 1959.) The extent, nature and interrelation of the solar and terrestrial atmospheres are considered.

551.510.52:523.74

Solar Activity and the Altitude of the Tropopause near the Equator-D. Stranz. (J. Atmos. Terr. Phys., vol. 16, pp. 180-182; October, 1959.) The height of the tropopause at Leopoldville and the Zurich sunspot number are shown to be correlated during 1953-1958. The consequent possibility of a relation between solar activity and tropospheric conditions is discussed.

551.510.53

Molecular Oxygen Densities in the Mesosphere at Fort Churchill-J. E. Kupperian, Jr., E. T. Byram and H. Friedman. (J. Atmos. Terr. Phys., vol. 16, pp. 174-178; October, 1959.) Densities between 70 and 86 km were lower in early spring by a factor 1.8 than in midsummer. Dissociation appeared to begin near 96 km in March, 1958 and 86 km in July, 1957.

551.510.535

A Contribution to the Theory of the Motion Weak Irregularities in the Ionosphereof P. C. Clemmow and M. A. Johnson. (J. Atmos. Terr. Phys., vol. 16, pp. 21-36; October, 1959.) If diffusion is neglected, any onedimensional irregularity travels unchanged with a constant velocity which depends on the magnitude and direction of the ionospheric electrostatic and magnetostatic fields. In the same way, a weak two-dimensional irregularity which is parallel to the direction of the magnetostatic field travels with a constant velocity and preserves its shape; more complicated irregularities do not preserve their shape.

551.510.535

A Discussion of the Motion in Nitrogen of Free Electrons with Small Energies with Reference to the Ionosphere-L. G. H. Huxley, (J. Atmos. Terr. Phys., vol. 16, pp. 46-58; October, 1959.) It is found that the collision cross section of electrons is proportional to their velocity. From this result, an accurate expression is derived for the velocity of electron drift in terms of the mean energy of the electrons. The energy losses are found to be mainly due to excitation of the rotational states of the nitrogen molecules. With these results, measurements of radiowave interaction can be used to deduce molecular densities in the height range 82-90 km. These densities are consistent with the ARDC model atmosphere.

551.510.535

A Theory of Electrostatic Fields in a Horizontally Stratified Ionosphere Subject to a Vertical Magnetic Field -D. J. Farley, Jr. (J. Geophys. Res., vol. 64, pp. 1225-1233; September, 1959.) A discussion of a possible explanation of spread-F and radiostar scintillation.

551.510.535

New Methods and Some Results Concerning True Ionospheric Height Calculations-W. Becker. (J. Atmos. Terr. Phys., vol. 16, pp. 67 83; October, 1959.) Three methods are described: a) an optical-graphical comparison method using given models, b) a general method applicable to monotonic h'(f) traces, and c) a method which applies to the model method corrections which are derived from differences between the actual and model h'(f)traces. It is shown how retardation in the E_s

layer can be used to estimate the depth of the minimum between the E and F layers.

551.510.535

Transient Fine Structure of the E Layer-W. Dieminger. (J. Atmos. Terr. Phys., vol. 16, p. 179; October, 1959.) A note on the fine structure of the E laver and its variability as observed using an ionosonde with high power and slow frequency variation.

551.510.535

The Ionospheric E Layer at Cape Hallett-G. A. M. King. (J. Atmos. Terr. Phys., vol. 16, pp. 186-187; October, 1959.) Ionograms were examined and the E-layer critical frequencies of those unaffected by ionization movements were compared with those expected on a simple Chapman model. Agreement was good when an electron recombination process was considered for the model.

551.510.535

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The Gyro-Frequency in the E-layer above Slough, England-W. R. Piggott. (J. Atmos. Terr. Phys., vol. 16, pp. 197-198; October, 1959.) "Measurements of the gyro-frequency, f_{II} in the *E* layer from the separation of f_0E and $f_x E$ at Slough give $f_H = 1.236 \pm 0.015$ mc. The calculated value is $f_{II} = 1.27$ mc. The difference is consistent with that found by Scott (2001 of 1951) using $f_0E - f_zE$ measured at high latitudes.

551.510.535 501 The Early-Morning E2 Layer and some Evidence of Pre-Sunrise F-Layer "Splitting"-P. Bandyopadhyay. (J. Atmos. Terr. Phys., vol. 16, pp. 84-92; October, 1959.) Observations at Haringhata, India, show that E2-layer cusps and ridges are regular sunrise phenomena at that location, with marked seasonal variations in character and frequency of occurrence. A "splitting" of the F layer during early morning in winter is also observed and the possible bearing of this on the E-layer phenomena observed is discussed.

551.510.535

502 Annual Distribution of Sporadic E-N. C. Gerson. (J. Almos. Terr. Phys., vol. 16, pp. 189-191; October, 1959.) The geographical distribution of E_{θ} ionization over North America during 1949 is studied from reports of VHF radio contacts.

551.510.535

Sporadic-E Ionization over Lindau/Harz during Last Year-W. Becker. (Arch. elekt. Übertragung, vol. 12, pp. 481-487; November, 1958.) Ionospheric sounding data obtained during the period August, 1957-July, 1958, are subjected to a detailed statistical analysis. More detailed investigations are proposed to ascertain the causes of E_s ionization.

551.510.535

Annual Wave in the World-Wide F-Region Ionization-B. N. Bhargava. (Indian J. Meteorol. Geophys., vol. 10, pp. 69-72; January, 1959.) An analysis of the noon median value for f_0F_2 at 31 stations indicates that the annual component R_1 varies with latitude in a manner very similar to that of the steady ionization R_0 . A similar analysis over a 9-year period for two of the stations gives a value of R_1 of the same order of magnitude as R_0 , with a maximum around the epoch of minimum sun-earth distance.

551.510.535

Note on the Cause of Ionization in the FRegion-M. H. Rees and W. A. Rense. (J. Geophys. Res., vol. 64, pp. 1251-1255; September, 1959.) In view of the high absolute intensity of the solar 303.8-Å He 11 line recently observed by rocket at 140 km and 212 km, electron densities at these levels were computed assuming that the 303.8-Å photons ionize oxygen atoms. These computed densities approximate closely to those measured by rocket at these heights.

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A New Theoretical Model of the Composite F Layer-F. Mariani. (J. Atmos. Terr. Phys., vol. 16, pp. 160-173; October, 1959.) It is suggested that bifurcation of the F layer is caused by a fairly rapid variation in temperature gradient and that tidal variations have only secondary effects. This model would explain the North-South asymmetry in electron density of the F_2 layer (see 3706 of 1959).

551.510.535

Single and Double Inflexions on the F-Trace-V. Marasigan. (J. Atmos. Terr. Phys., vol. 16, pp. 193-196; October, 1959.) An attempt is made to explain the bifurcation of the daytime F layer in terms of the Appleton-Lyon theory of the "height lag" (Appleton and Lyon, Physics of the Ionosphere, pp. 20-39; 1955). Flayer models are considered and, since the height lag decreases through the day, bifurcation takes place.

551.510.535

The Effect of the F_1 Layer on the Calcula-tion of the Height of the F_2 Layer—M. D. Vickers. (J. Atmos. Terr. Phys., vol. 16, pp. 103–105; October, 1959.) "An approximate relationship between the estimated true height of the peak of the F_2 layer, and that given by assuming a single layer having a parabolic electron-density distribution, is derived and compared with experimental data."

551.510.535:523.78

Anomalous Ionospheric Reflection during Solar Eclipses-W. L. Price. (J. Atmos. Terr. Phys., vol. 16, pp. 93–98; October, 1959.) Analysis of electron densities during an eclipse shows that while the slope and curvature of layer strata are mostly very small, effective discontinuities can occur which would produce complexities in ionograms These complexities are due to rays reflected along paths inclined to the vertical.

551.510.535:523.78

Ionospheric Observations on the F Region during the Solar Eclipse of 19 April 1958-S. Datta, P. Bandyopadhyay and R. N. Datta. (J. Atmos. Terr. Phys., vol. 16, pp. 182–185; October, 1959.) Results obtained at Haringhata, India, demonstrate eclipse effects in the F_2 layer for maximum electron density, total electron content of a unit column and true height of the layer peak.

551.510.535:539.16

On Artificial Geomagnetic and Ionospheric Storms associated with High-Altitude Explosions-S. Matsushita. (J. Geophys. Res., vol. 64, pp. 1149-1161; September, 1959.) Associated with the nuclear explosions at Johnston Island on August 1 and 12, 1958, were a) circular electric currents, explicable on the hypothesis of the dynamo effect, which caused observed magnetic variations at places up to 2200 km distant, b) fast particles traveling along magnetic lines of force and causing aurora and magnetic storms at Apia, c) X rays which increased D-region ionization by about eight times at Maui.

551.510.535:551.507.362.2 512

The Electron Density in the Outer Ionosphere-L. Klinker, R. Knuth and K. H. Schmelovsky. (Z. Meteorol., vol. 13, pp. 192-193; July/August, 1959.) Faraday fading records at Kühlungsborn (East Germany) of 20- and 40-mc transmissions from satellite 195882 during daytime transits in summer

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the Electronic Component of Cosmic Rays-H. Tuniner, (Monthly Notices Royal Astron. Soc., vol. 119, no. 2, pp. 184-193; 1959.) The similarity between the spectra of cosmic rays and of the relativistic electrons responsible for cosmic radio waves suggests that the electrons may derive from the collisions of primary cosmic rays with the interstellar gas. The relation to be expected from this process is examined.

523.164.3

The Brightness Distribution within the Radio Sources Cygnus A (19N4A) and Cassiopeia A (23N5A)-R. C. Jennison and V. Latham. (Monthly Notices Royal Astron. Soc., vol. 119, no. 2, pp. 174-183; 1959.) A description of the results obtained with a three-station interferometer system operating at 127 mc.

523.164.3 468 The Source of Radiation from Jupiter at Decimetre Wavelengths-G. B. Field. (J. Geophys. Res., vol. 64, pp. 1169-1177; September, 1959.) Electrons from the sun which are trapped in Jupiter's magnetic field may be the source.

523.164.32

Radio Emission in the Outer Corona-W. C. Erickson. (Phys. Rev. Lett., vol. 3, pp. 365-367; October 15, 1959.) During May, June and July transit observations of the sun were obtained at a frequency of 26.3 mc at Clark Lake radio-astronomy station. The height of the radio emission appears to have been 4-5 solar radii early in May.

523.164.32:523.746 470 Correlation between the Intensity of the Umbra of Sunspots and Enhanced Radiation on 200 Mc/s-P. Maltby, (Nature, vol. 184, suppl. no. 18, p. 1391; October 31, 1959.) From observations made in Norway since April, 1959, a high correlation is shown to exist between the darkness of the umbra and the noise activity of sunspots.

523.164.32:550.385.4 471 Geomagnetic Disturbance and Velocity of Slow-Drift Solar Radio Bursts-M. B. Wood and C. S. Warwick. (Nature, vol. 184, suppl. no. 19, pp. 1471-1472; November 7, 1959.) Frequency drift rates of type-II bursts, determined from radio spectral observations, indicate a systematically greater acceleration for the sources of bursts which are followed by geomagnetic disturbance.

523.164.4:535.221 472 A Radio-Astronomical Test of the Ballistic Theory of Light Emission-L. R. O. Storey and R. S. Lawrence, (Observatory, vol. 79, pp. 150-151; August, 1959.) A comparison between optical and radio-interferometer measurements of the declination of radio star Cygnus-A is put forward as proof that the ballistic theory of light emission [see, e.g., Monthly Notices Royal Astron. Soc., vol. 119, no. 1, pp. 67–71; 1959 (Dingle)] is untenable.

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473 On the Possibility of Detecting Synchrotron Radiation from Electrons in the Van Allen Belts-R. B. Dyce and M. P. Nakada. (J. Geophys. Res., vol. 64, pp. 1163-1168; September, 1959.) The equipment suggested is a 30me polarimeter at the magnetic equator, using an antenna directed vertically upward.

523.165

Proton Component of the Primary Cosmic Radiation-F. B. McDonald and W. R. Webber. (Phys. Rev., vol. 115, pp. 194-205; July 1, 1959.) The proton component has been studied at high altitudes on a series of balloon flights at

various latitudes using the Čerenkov scintillation-counter technique. Results are discussed.

523,165 475 Unusual Cosmic-Ray Fluctuations on July 17 and 18, 1959-H. Carmichael and J. F. Steljes. (Phys. Rev. Lett., vol. 3, pp. 392-394; October 15, 1959.) A large Forbush decrease of cosmic-ray intensity which coincided with a magnetic storm exhibited rapid changes of neutron intensity, at the rate of 7 per cent in 20 minutes. It is difficult to account for these changes on the basis of existing theories of the modulation of cosmic radiation.

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North-South Anisotropy and Anticipatory Increase of Intensity associated with the Cosmic-Ray Storm of February 11, 1958-V. Sarabhai and R. Palmeira. (Nature, vol. 184, pp. 1204-1207; October 17, 1959.) An analysis of cosmic-ray data obtained from a highcounting-rate meson detector and a grid of neutron monitor stations during the period February 9-12, 1958. Other effects associated with the storm are discussed.

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523.165:523.75 477 Observations of Low-Energy Solar Cosmic Rays from the Flare of 22 August 1958-K. A. Anderson, R. Arnoldy, R. Hoffman, L. Peterson and J. R. Winckler. (J. Geophys. Res., vol. 64, pp. 1133-1147; September, 1959.)

523.165:523.75 478 Observations of Solar Flare Radiation at High Latitude during the Period July 10-17, 1959-R. R. Brown and R. G. D'Arcy, (Phys. Rev. Lett., vol. 3, pp. 390-392; October 15, 1959.) Cosmic-ray detectors, consisting of photon counters, were borne aloft by clusters of sounding balloons. Curves showing variations of intensity with atmospheric depth are given for quiet conditions and under solar-flare conditions. The results are analyzed.

523.42:621.396.96 479 Radio Echo Observations of Venus-J. V. Evans and G. N. Taylor, (Nature, vol. 184, pp. 1358-1359; October 31, 1959.) A description of the equipment used and a discussion of the results of observations at Jodrell Bank on 408 mc during September, 1959. No echoes stronger than the noise level of the receiver were observed; an analysis of the signals received was therefore made with integrating equipment and the results compared with those of Price, et al. (2556 of 1959).

523.5:621.396.9 480 The Effect of Trail Irregularities on the Interpretation of Meteor Echoes-A. G. Mc-Namara and D. W. R. McKinley, (J. Atmos. Terr. Phys., vol. 16, pp. 156-159; October, 1959.) "A brief discussion of some points raised in a recent paper by Manning (2225 of 1959) on obliquely-scattered meteor echoes is followed by the suggestion that the initial distribution of ionization along a typical meteor trail is markedly irregular. Several tentative hypotheses are advanced to account for the irregularities.

523. 755: 523.164 481 The Inner Solar Corona during June 1959-G. A. Newkirk, G. W. Curtis, D. K. Watson, R. Manning and J. Shelby, (Nature, vol. 184, suppl. no. 17, pp. 1308-1309; October 24, 1959.) An analysis of observations with the K-coronameter at Climax, Colorado, of the Taurus-A radio source.

550.385:523.78

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On the Variation in the Horizontal Intensity of the Geomagnetic Field at Phalodi (Rajasthan) during the Solar Eclipse of 30 June

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1954-B. J. Srivastava and N. S. Sastri. (Indian J. Meteorol. Geophys., vol. 10, pp. 73-84; January, 1959.) A detailed study of horizontal field variations during the period June 22 to July 7, 1954, indicates that a fall of 10γ 16 minutes after totality may not be an eclipse effect. The results therefore do not confirm those of Egedal and Ambolt (107 of 1956).

550.385.2 483 Variations in the Geomagnetic Field at Ibadan, Nigeria: Parts 1 and 2-C. A. Onwumechilli and N. S. Alexander. (J. Atmos. Terr. Phys., vol. 16, pp. 106-123; October, 1959.) Magnetic records have been analyzed for the period November, 1955-June, 1957 to detect solar and lunar terms in the variations of the usual magnetic components. Amplitude variations of H and Z for both terms are about three times larger than for stations which have the same geographic latitude but which are not near the magnetic equator. These large variations are attributed to the electrojet. The solar variations of D show no such obvious effect.

551.507.362.2 484 Effects of the Earth's Oblateness on the Orbit of an Artificial Satellite-A. de Moraes. (Ann. acad. brasil. sci., vol. 30, pp. 465-510; December 31, 1958. In English.) A mathematical analysis of the perturbations of the radius vector, the displacement of the modal and

apsidal lines and the variation of the inclination with latitude. A rigorous solution to the second-order differential equation is given for an equatorial orbit taking account of the firstorder effects of the earth oblateness. Some second-order effects on the equatorial orbit and the relativistic effect of the apsidal-line displacement are considered.

551.507.362.2

Density of the Upper Atmosphere from Analysis of Satellite Orbits : Further Results-D. G. King-Hele. (Nature, vol. 184, pp. 1267-1270; October 24, 1959.) A method described earlier (2955 of 1959) has been refined to take into account atmospheric rotation.

551.507.362.2:551.510.535 486 The Ion-Trap Results in "Exploration of the Upper Atmosphere with the Help of the Third Soviet Sputnik"—E. C. Whipple, Jr. (PROC. 1RE, vol. 47, pp. 2023-2024; November, 1959.) The interpretation of the data given by Krassovsky (1550 of 1959) is queried, and an analysis presented which leads to lower values of vehicle potential and electron temperature.

551.507.362.2:621.391.812 487

Unusual Propagation of Satellite Signals-M. Dewan. (PROC. IRE, vol. 47, p. 2020; November, 1959.) Gives example of signal strength maxima for Sputnik I.

551.507.362.2:621.391.812.63 488 High-Frequency Fading Observed on the 40 Mc/s Wave Radiation from Artificial Satellite 1957 α -G. S. Kent. (J. Atmos. Terr. Phys., vol. 16, pp. 10-20; October, 1959.) This fading is attributed to irregularities in the Fregion of the ionosphere. The properties of these irregularities are examined and compared with those thought to be responsible for radio star scintillations and spread-F echoes.

551.507.362.2:621.396.41 480 Multiplexing Techniques for Satellite Applications-King. (See 664.)

551.508.822:551.594 490 Radiosondes for Measurements of Atmospheric Electricity-R. Mühleisen and H. L Fischer. (Arch. tech. Messen, no. 274, pp. 229

circuit parameters for the three basic transistor configurations are tabulated. The main types of feedback and their effects are discussed.

621.375.9:537.56:538.56 437 Possible Low-Noise Electron-Beam Plasma Amplifier—Anderson. (See 455.)

621.375.9: 538.569.4.029.64 438 Cavity Maser Experiments using Ruby at S-Band-W. S. C. Chang, J. Cromack and A. E. Siegman. (J. Electronics Control, vol. 6, pp. 508-526; June, 1959.) A three-level solidstate maser was used at frequencies near 3 kmc. Pumping at 13.5 kmc gave a gain-bandwidth product of over 50 mc. Another mode of operation with pumping at 23.7 kmc is capable of increasing this product substantially.

621.375.9:538.569.4.029.64 439 A Double Pumping Scheme Applicable to Low-Frequency Masers-J. E. King, A. Birko and G. Makhov. (Proc. IRE, vol. 47, p. 2025; November, 1959.) Note on a "parallel" pumping system of importance for low-frequency operation of a ruby maser.

621.375.9:621.382.2 440 Semiconductor Varactors using Surface Space-Charge Layers-Pfann and Garrett. See (689.)

621.375.9:621.382.2 441 Alloyed, Thin-Base Diode Capacitors for Parametric Amplification-K. E. Mortenson. (J. Appl. Phys., vol. 30, pp. 1542-1548; October, 1959.) The design principles are given; maximum Q-values of the order 200 at 1 kmc have been obtained with capacitive swings greater than 10.

621.375.9:621.383:535.376 442 Electroluminescent Cell Applications-R. B. Lochinger and M. I. O. Strutt, (Electronic Radio Eng., vol. 36, pp. 398-406; November, 1959.) An account of investigations of the combination of electroluminescent cells and photoreceptors as elements in amplifier, oscillator, demodulation-amplifier, and bistable multivibrator circuits. Measurements of the variation of efficiency with frequency and a theoretical analysis of photoconductor amplifier time-constants are also given.

621.376.4

Diode Phase-Sensitive Detectors with Load-R. Chidambaram and S. Krishnan. (Electronic Engrg., vol. 31, pp. 613-616; October, 1959.) A theoretical and experimental investigation of the nonlinearity introduced by loading a simple diode push-pull phase detector

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530.12:531.18:621.3.018.41(083.74) 444 A New Experimental Test of Special Rela-

tivity-J. P. Cedarholm and C. H. Townes. (Nature, vol. 184, pp. 1350-1351; October 31, 1959.) The experiment is based on comparing the frequencies of two NH3-beam masers mounted with oppositely directed beams on a rack which may be rotated through 180° about a vertical axis. A precision of one part in 1012 has been achieved in this frequency comparison, from which the upper limit on an ether drift of 1/1000 of the earth's orbital velocity may be set.

535.62

Two-Coordinate Colour-M. H. Wilson and R. W. Brocklebank. (Electronic Radio Eng., vol. 36, p. 429; November, 1959.) Com-ment on 3629 of 1959 with reference to other work on the subject.

537.311.1

On the Screening of Impurity Potential by

Conduction Electrons-N. Takimoto. (J)Phys. Soc. Japan, vol. 14, pp. 1142-1158; September, 1959.) A modified Thomas-Fermi method is used to calculate the impurity potential, and it is shown that similar results are obtained by the method of Nakajima and Bardeen and Pines (see 379 of 1956),

537.525:621.372.413

The Effect of Field Configuration on Gas Discharge Breakdown in Microwave Cavities at Low Pressure-S. A. Self and H. A. H. Boot. (J. Electronics Control, vol. 6, pp. 527-547; June, 1959.) A new regime of gas discharge breakdown in microwave cavities is shown to be due to gradients in the electric field amplitude. Experimental results are given.

537.533 448 Necessary and Sufficient Trajectory Conditions for Dense Electron Beams-W. M. Mueller (J. Flectronics Control, vol. 6, pp. 499-507; June, 1959.) Gives conditions for flow in the direction of one coordinate in a number of coordinate systems.

537.533

Two Alternative Definitions of Small-Signal RF Power of Electron Beams-E. L. Chu. (J. Appl. Phys., vol. 30, pp. 1617-1618; October, 1959.) Lagrangian or Eulerian definition is chosen according to the type of problem or method of calculation.

537.533 450 Comments on Klüver's Paper entitled "Small-Signal Power Conservation Theorem for Irrotational Electron Beams"-E. L. Chu. (*J. Appl. Phys.*, vol. 30, pp. 1618–1619; October, 1959.) See 2703 of 1958.

537.56

451 Measurement of Plasma Temperature and Electron Density-K. Murakawa and S. Mizuno Hashimoto. (J. Phys. Soc. Japan, vol. 14, pp. 1235-1242; September, 1959.) The plasma temperature and electron density were obtained from a comparison of the wavelength of the line NE I λ 5852 emitted from an arc discharge and from a condensed spark discharge.

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Electron and Ion Runaway in a Fully Ionized Gas: Part 1-H. Dreicer. (Phys. Rev., vol. 115, pp. 238-249; July 15, 1959.) Hydrodynamic equations are used to describe the flow of the electrons and ions of a fully ionized gas under the action of an electric field of arbitrary magnitude.

537.56:061.3

Ionization Phenomena in Gases-J. Dutton, D. Harcombe and E. Jones. (Nature, vol. 184, pp. 1353-1358; October 31, 1959.) Report of a conference at Uppsala, Sweden, August 17, 1959

537.56:538.56

454 Effect of Relatively Strong Fields on the Propagation of E. M. Waves, through a Hypersonically Produced Plasma-W. B. Sisco and J. M. Fiskin. (IRE TRANS. ON AN-TENNAS AND PROPAGATION, vol. AP-7, pp. 240-244; July, 1959. Abstract, PRoc. IRE, vol. 47, p. 2037; November, 1959.)

537.56:538.56:621.375.9

Possible Low-Noise Electron-Beam Plasma Amplifier-J. M. Anderson. (J. Appl. Phys., vol. 30, pp. 1624-1625; October, 1959.) Lownoise amplification of space-charge waves should be achieved by interaction between electron beam and plasma in the negative-glow region of a cold-cathode discharge.

537.56:538.69

Plasma Configurations with Surface Cur-

rents which are Held in Equ librium by a Magnetic Field-R. Kippenhahn. (Z. Naturforsch., vol. 13a, pp. 260-267; April, 1958.) The conditions for the existence of various plasma configurations in equilibrium with an external magnetic field are investigated.

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Remarks on Magnetically Dilute Systems-H. Sato, A. Arrott and R. Kikuchi. (J. Phys. Chem. Solids, vol. 10, pp. 19-34; April, 1959.) A re-examination of the problem using an Ising model

538.3 A Scalar Representation of Electromagnetic Fields: Part 3-P. Roman. (Proc. Phys.

Soc., vol. 74, pp. 281-289; September 1, 1959.) Gives transformation properties and the physical energy-momentum tensor appropriate to Green and Wolf's theory (1739 of 1954).

538.3 459 A Scalar Representation of Electromagnetic Fields: Part 2-E. Wolf. (Proc. Phys. Soc., vol. 74, pp. 269-280; September 1, 1959.) Extends work in an earlier paper [1739 of 1954 (Green and Wolf)] to deal with energy transport.

538.566:535.42

Diffraction of a Dipole Field by a Unidirectionally Conducting Semi-Infinite Screen-J. Radlow. (Quart. Appl. Math., vol. 17, pp. 113-127; July, 1959.) An exact solution of the diffraction problem for a dipole is obtained.

538.566:535.42 461 Diffraction of Electromagnetic Waves in a

Band of Finite Width-G. A. Grinberg. (Dokl. Akad. Nauk SSSR, vol. 129, pp. 295–298; November 11, 1959.) Brief mathematical analysis based on a new method for solving integral equations similar to the Fredholm equations. An asymptotic form of the required solution for $\gamma > 1$ is obtained, where γ is a function of the number of waves and of the bandwidth.

538.569:539.2

Induced and Spontaneous Emission in a Coherent Field: Part 2-I. R. Senitzky. (Phys. Rev., vol. 115, pp. 227-237; July 15, 1959.) "The interaction between the electromagnetic field and a number of identical atomic systems, individually characterized by an electric dipole moment and two energy levels, is analyzed for the case where the atomic systems are inside a lossy cavity and exposed to a coherent driving field, resonance being assumed between atomic system, cavity, and driving field." Part 1-3777 of 1958.

538.569.4.029.6:535.343.9:537.228.5 463 100-kc/s Square-Wave Modulator and Re-

ceiver for Stark-Effect Microwave Spectrometers-H. G. Fitzky. (Z. angew. Phys., vol. 10, pp. 489-495; November, 1958.)

539.2:538.1

Superexchange Interaction and Symmetry Properties of Electron Orbitals-J. Kanamori, (J. Phys. Chem. Solids, vol. 10, pp. 87-98; July, 1959.)

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523.164 465 A Search for Neutral Atomic Hydrogen in Globular Clusters-M. S. Roberts. (Nature, vol. 184, suppl. no. 20, pp. 1555-1556; November 14, 1959.) Results of measurements of the 21-cm emission of two globular clusters M3

523.164:523.165 466

and M13 are discussed.

The Relation of Cosmic Radio Emission to

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together with a preliminary evaluation of their performance compared with existing techniques.

621.318.57:621.372.44 410 Signal Converter by Magnetic Cores for Parametron Device—K. Hanawa and K. Kusunoki. (Rep. elect. Commun. Lab., Japan, vol. 7, pp. 25-31; February, 1959.) Input signals for a parametron device [see 3588 of 1959 (Goto)] may be switched using ferromagnetic cores whose permeability is controlled by direct voltage.

621.318.57:621.382.3 411 Transistor Switching Speed-P. M. Thompson and J. Bateson. (Wireless World, vol. 65, pp. 530-533; December, 1959.) Theo-

retical treatment of the limitations of transistors in high-speed switches, together with some methods of improving performance by decreasing switching time.

621.319.4:537.529 412 The Breakdown Strength of Capacitors-II. Veith. (Frequenz, vol. 12, pp. 348-353; November, 1958.) A formula for breakdown field strength is derived which contains the coefficient of thermal conductivity, the dielectric constant, and a factor giving carrier mobility.

621.319.4.004.6 413 The Temperature and Voltage Dependence of the Length of Life of Capacitors-II. Veith. (Frequenz, vol. 12, pp. 353-355; November, 1958.) The process of deterioration of the dielectric material under various operating conditions is discussed on a theoretical basis (see 412 above) and with reference to life tests on paper capacitors.

621.372.5 414 Contribution to the Theory of General Quadripoles-O. Heymann. (Arch. elekt. Übertragung, vol. 12, pp. 488-496; November, 1958.) The matrix characteristics of general linear quadripoles are investigated, and a strict definition of a lossless quadripole is given. An equivalent circuit of the quadripole formed by three partial quadripoles connected in tandem is derived; matching problems and a formula for power gain are discussed.

415 621.372.54 Design of Attenuators of Given Characteristics-U. Kirschner. (Elektron. Rundschau, vol. 12, pp. 412-414; December, 1958.) Design formulas for quadripole attenuating networks are tabulated.

621.372.54 416 A Recriprocal Theorem on Quasilinear Wave Filters-A. W. Thies. (J. Inst. Eng. (Australia), vol. 31, pp. 243-246; October/November, 1959.) Under certain conditions, the power of an intermodulation product measured at one pair of terminals of a filter equals that measured at the other pair when all signals are transmitted in the opposite direction but at unchanged levels.

621.372.54:621.373.1 417 Electromechanical Filters for Use in Telecommunication Equipment-G. L. Grisdale. (Brit. Commun. Electronics, vol. 6, pp. 768-772; November, 1959.) Includes descriptions of construction and performance of reed-type, magnetostrictive and piezoelectric filters.

621.372.543.2:538.652 418 A Practical Electromechanical Filter-H. Bache. (Marconi Rev., vol. 22, pp. 145-153; 3rd Quarter, 1959.) Details of materials and fabrication techniques are given for narrow-band 1 mw torsional filters in the 100-500-kc range. 621.372.543.2:538.652

A Theoretical Analysis of the Torsional Electromechanical Filters-W. Struszynski. (*Marconi Rev.*, vol. 22, pp. 119–143; 3rd Quarter, 1959.) The mechanical properties of a torsional system are expressed in terms of electrical equivalents. By introducing a "trans-ducer transfer ratio" with the dimensions of charge, a method is developed for the design of electromechanical filters based on equivalent electrical networks. Pass band ripple, spurious modes and transducer matching are discussed.

621.373.018.41-52:621.374.32 420 Digital Input for Precision Variable Oscillators-N. G. Alexakis. (Electronics, vol. 32, pp. 56-57; October 30, 1959.) Details are given of a signal generator for the frequency range 1 cps-1 mc. The drive unit is a voltage-controlled oscillator regulated by the difference in width between standard 1-second pulses and pulses whose duration is that of the time required for the oscillations fed to a counter unit to equal a preset number. Successive sampling at intervals of 1.11 second ensures errors less than 0.01 per cent.

621.373.2

A Method of Generating Pairs of Millimicrosecond Current Pulses Separated by a Variable Interval-I. M. Somerville. (Proc. Phys. Soc., vol. 74, pp. 378-379; September 1, 1959.) The first pulse is formed by the discharge of a coaxial line L through a spark gap into a variable length section of identical line L_1 . A second pulse is produced by the return of the first pulse after reflection at the shortcircuited end of L_1 . Further reflections from the other end of L are prevented by a clipping tube.

621.373.42 422 Frequency-Stable Oscillators for Current and Voltage-W. Herzog. (Nachrichtentech. Z., vol. 11, pp. 550-556; November, 1958.) The suitability of bridge-type oscillators as current or voltage source under optimum frequencystability conditions is discussed.

621.373.421

Wien-Bridge Oscillators-D. E. D. Hickman. (Wireless World, vol. 65, pp. 550-555; December, 1959.) Theoretical analysis of Wienbridge oscillators and procedure for practical design, together with an example of thermistor stabilization.

621.373.443 424 Pulse Modulators using Transistors and Switching Reactors-B. F. C. Cooper and W. J. Payten. (Proc. IRE (Australia), vol. 20, pp. 148-152; March, 1959.) A regenerative circuit is described which uses a power transistor to control the charging of a pulse-forming network through a transformer with a sharply saturating core.

621.373.52 425 Point-Contact Transistor Relaxation Oscillators-V. N. lakovlev. (Radiotekh. Elektron., vol. 3, pp. 61-73; January, 1958.) The plotting of phase curves for oscillators with emittercollector and collector capacitance are shown and possible operating modes investigated. Expressions are given for deriving the pulse parameters. Conditions are expressed for the appearance of step discontinuities as well as for

a self-oscillatory mode.

621.374.3:621.3.018.7 426 Approximate Waveform Solutions for Diodes in Pulse Circuits-D. C. Dillistone. (Electronic Engrg., vol. 31, pp. 607-610; October, 1959.) With an approximate representation of a diode, solutions are obtained for the response of simple circuits to a rectangularwave input. Applications of the results are discussed.

621.374.32:621.376

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427 An Investigation into some Aspects of Diode Quantizing Circuits-II. V. Bell and W. Alexander, (Electronic Engrg., vol. 31, pp. 594-598; October, 1959.) Quantization is defined and work in the field is reviewed. Three circuits are compared theoretically and by measurement, and possible applications are described.

621.374.4:621.373.3.029.64 428

Harmonic Generation in a Cyclotron Resonant Plasma-R. M. Hill and S. I. Tetenbaum. (J. Appl. Phys., vol. 30, pp. 1610-1611; October, 1959.) The harmonic conversion efficiencies are comparable to those for crystals and. over the same power range, superior to those for ferrite multipliers.

621.374.44:621.382.3 429

A Transistor Blocking-Oscillator Frequency Divider-F. Butler. (Electronic Engrg., vol. 31, pp. 611-612; October, 1959.) This includes the design of a "staircase" waveform generator in which all the voltage increments are equal in amplitude.

621.375.018.75

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430 Design of Pulse Amplifier-R. C. Ganguli. (Indian J. Phys., vol. 33, pp. 263-275; June, 1959.) A relation between gain, overshoot and risetime is derived and applied in the design of a single-stage tube amplifier.

621.375.2.018.7

Distortion in Pentode Voltage Amplifiers-R. E. Aitchison, C. T. Murray and I. S. Docherty. (Proc. IRE (Australia), vol. 20, pp. 147-148; March, 1959.) Characteristics are given which show that, for a fixed screen voltage, the distortion varies rapidly with changes in grid bias, while, if the screen voltage is supplied via a series resistance, there is a compensating action which maintains the distortion at an almost constant value.

621.375.223

RC Amplifier with 60-mc/s Bandwidth-K. J. Schmidt-Tiedemann. (Elektron. Rundschau, vol. 12, pp. 414-416; December, 1958.) The effect of stray capacitance is compensated by a cathode-follower circuit. In the two-stage circuit described a gain of 5.3 is achieved for a bandwidth of 62 mc.

621.375.227 433

Cathode-Coupled Push-Pull Output Stage K. R. Sturley and J. P. Bennett. (Electronic Radio Eng., vol. 36, pp. 410-415; November, 1959.) A theoretical investigation of linear operation shows that for large commoncathode resistances, R_k , the ratio of anode currents in each tube approximates to unity and their magnitude is almost independent of R_k . This is confirmed experimentally, and measurements of power output and distortion are obtained for different values of R_k .

621.375.3

How Magnetic Amplifier controls Transconductance—C. C. Whitehead. (*Electronics*, vol. 32, pp. 84–87; November 13, 1959.) See 2170 of 1959.

621.375.4:621.396 435

Transistor Amplifiers for Sound Broad-casting—S. D. Berry. (B.B.C. Engrg. Monographs, no. 26, 19 pp.; August, 1959.) The application of Ge p-n-p junction transistors to various types of high-quality amplifiers is described. An assessment is made of the suitability of transistors for this field.

621.375.4.029.4/5

Transistors in Low-Frequency Amplifiers-W. Langsdorff and W. Heberle. (Frequenz, vol. 12, pp. 337-348; November, 1958.) Design formulas are derived and the relations between
Z., vol. 11, pp. 561-564; November, 1958.) The effects of misalignment, such as offset and twist, between sections of rectangular waveguide and of discontinuities in cross section due to manufacturing tolerances are discussed. The maximum reflections possible under German and U. S. manufacturing standards are estimated.

621.372.832.8

E-Type X Circulator-S. Yoshida. (PROC. IRE, vol. 47, pp. 2017-2018; November, 1959.) The circulator comprises a ferrite element in an E-type four-port rectangular waveguide junction.

621.372.832.8 An E-Type T Circulator-S. Voshida. (PROC. IRE, vol. 47, p. 2018; November, 1959.)

The device has a ferrite element in an ordinary *E*-type *T* junction.

621.372.85

The Effect of a Dielectric Film on the Attenuation of the H_{01} Wave in a Rectilinear Quasicircular Waveguide-B. Z. Katsenelenbaum. (Radiotekh. Elektron., vol. 3, pp. 38-45; January, 1958.) Calculation of the additional attenuation of H_{0m} waves produced by a thin semiconducting film located on the inside surface of a circular waveguide. The analysis of waveguides of almost circular cross section is generalized to that of irregular rectilinear waveguides.

621.372.85 386 Symmetrical Diaphragm of Arbitrary Thickness in a Circular Waveguide-M. V. Butrov. (Radiotekh. Elektron., vol. 3, pp. 56-60; January, 1958.) A mathematical analysis of the effect of the round aperture of the diaphragm on the passage of H_{01} -type waves.

621.372.85

387 Propagation of Electromagnetic Waves in Loaded Bent Waveguides-A. N. Didenko. (Radiotekh. Elektron., vol. 4, pp. 172-180; February, 1959.) Dispersion equations and field expressions are derived and various waveguide systems are considered in relation to their application in particle accelerators.

621.372.852.22 388 A Ferrite Boundary-Value Problem in a Rectangular Waveguide—L. Lewin. (Proc. IEE, Part B, vol. 106, pp. 559-563; November, 1959.) A solution is given for the reflection of an electromagnetic wave from a transversely magnetized ferrite block in a rectangular

621.396.67:537.226

A Variational Expression for the Terminal Admittance of a Semi-Infinite Dielectric Rod-C. M. Angulo and W. S. C. Chang. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 207-212; July, 1959. Abstract, PRoc. IRE, vol. 47, pp. 2036-2037; November, 1959.)

621.396.677

waveguide.

Optimum Linear Co-Phased Aerials with Continuous Current Distribution-I. F. Sokolov and D. E. Bakman. (Radiotekh. Elektron., vol. 3, pp. 46-55; January, 1958.) A mathematical analysis based on Dolph's investigation of current distribution for broadside arrays (2487 of 1946) which optimizes the relation between beamwidth and size of the sidelobes. Formulas for the determination of directivity diagrams and the current amplitude distribution are derived, and optimum and quasi-optimum directivity diagrams for different sidelobes are shown.

621.396.677.029.55

A Multiple-Direction Universally-Steerable Aerial System for H.F. Operation-D. W. Morris and G. Mitchell, (Proc. IEE, Part B. vol. 106, pp. 555-558; November, 1959.) The antenna can be steered in both azimuth and elevation. It comprises a number of omnidirectional unit antennas with the outputs (in the receiving condition) arranged to be in phase for any desired combination of frequency and direction.

621.396.677.3

383

384

385

392 On the Phase Velocity of Wave Propagation along an Infinite Yagi Structure-D. L. Sengupta. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 234-239; July, 1959. Abstract, PRoc. IRE, vol. 47, p. 2037; November, 1959.)

621.396.677.3:621.396.965

An Investigation of the Complex Mutual Impedance between Short Helical Array Elements-A. R. Stratoti and E. J. Wilkinson. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 279-280; July, 1959.)

621.396.677.7:621.372.826

A Note on Surface Waves along Corrugated Structures-L. O. Goldstone and A. A. Oliner. (IRE TRANS. ON ANTENNAS AND PROPAGAtion, vol. AP-7, pp. 274-276; July, 1959.) Comment on a paper by Hougardy and Hansen (3201 of 1959).

621.396.677.7:621.372.826 305 Comments on "Scanning Surface-Wave Antennas-Oblique Surface Waves over a Corrugated Conductor"-R. E. Collin: R. W. Hougardy and R. C. Hansen. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 276-277; July, 1959.) See 394 above.

621.396.677.71

396 Radiation from Slot Arrays on Cones-R. F. Goodrich, R. E. Kleinman, A. L. Maffett, C. E. Schensted, K. M. Siegel, M. G. Chernin, H. E. Shanks and R. E. Plummer. (IRE TRANS, ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 213-222; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.396.677.833:621.396.965

397 A Study of Spherical Reflectors as Wide-Angle Scanning Antennas-T. Li. IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 223–226; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.396.677.833.1

Analysis and Reduction of Scattering from the Feed of a Cheese Antenna-W. A. Cumming, C. P. Wang and S. C. Loh. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7. pp. 226-233; July, 1959. Abstract, PROC. IRE, vol. 47, p. 2037; November, 1959.)

621.396.677.85

A Method to Achieve a Collimated Circularly Polarized Beam-C. L. Gray and J. C. Huber, Jr. (IRE TRANS. ON ANTENNAS AND PROPAGATION, vol. AP-7, pp. 281-282; July, 1959.) Description of a lens which may be used with any linearly polarized source such as a waveguide horn.

AUTOMATIC COMPUTERS

681.142

380

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301

The Simulation of Equations with Analogue Computers-H. Schuchardt. VDI Zeitschrift, vol. 101, pp. 1053-1063; August 1, 1959.) Detailed description of the principles of operation of the various computer elements giving examples of programming and computation with reference to an American electronic analog computer.

681.142:517.9

Some Aspects of the Logical and Circuit Design of a Digital Field Computer-I. F.

World Radio History

Brown and B. Meltzer. (Electronic Engrg., vol. 31, pp. 590-592; October, 1959.) A new type of digital computer for the solution of field problems is described. By making calculations at all the lattice points of the field simultaneously, computation time is greatly reduced. An experimental design of a basic unit for potential and other problems is presented. See 3354 of 1958 (Meltzer and Brown).

681.142:621.318.042

393

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The AC Writing Method for Magnetic Core Matrices-S. Yamada and T. Bessho. (Rep. elect. Commun. Lah., Japan. vol. 7, pp. 44-47; February, 1959.) Description of a writing method suitable for operation in a system using parametron devices. Two alternating currents of frequency f and f/2 respectively produce an asymmetrical field in a toroidal core. One of the writing currents is also used for reading. See also 410 below.

681.142:621.318.57

Function Generation with Operational Amplifiers—H. Koerner and G. A. Korn. (Electronics, vol. 32, pp. 66-68, 70; November 6, 1959.) Errors in analog computers caused by diode limiters can be reduced by using accurate electronic switches with high-gain dc amplifiers and voltage feedback. Applications of the techniques for comparators, multivibrators and integrator reset circuits are described and future applications are indicated.

681.142:621.372.44

Reading of Recorded Signals with a Low-Frequency Parametron-K. Zen'iti and K. Nisiguit. (Rep. elect. Commun. Lab., Japan, vol. 7, pp. 48-53; February, 1959.) Description of a tape recording system in which parametron devices replace tubes. To increase the storage capacity, some form of multiplex system is required.

681.142:621.395.625.3 405

High-Density Recording on Magnetic Tape-A. Gabor. (Electronics, vol. 32, pp. 72-75; October 16, 1959.) A self-clocking technique is used to by-pass problems of digital recording at a density of 1500 bits. High reliability is obtained with no information drop-out.

681.142:681.188

The Application of a Pattern-Recognition Technique to the Synthesis of Coding Circuits -J. H. Calderwood and A. Porter. (J. Electronics Control, vol. 6, pp. 556-566; June, 1959.) If the coding processes considered are regular and predictable, then corresponding translating circuits can be determined. To illustrate the technique a pattern-recognition method is applied to two coding problems.

681.142:681.188

The Synthesis of a Parallel Adder Circuit using a Pattern-Recognition Technique-J. H. Calderwood and A. Porter. (J. Electronics Control, vol. 6, pp. 567-576; June, 1959.) Besides the synthesis, a comparison with a conventional adder circuit is given. See 406 above.

681.142:681.42.002.2 408

Lens Designing by Electronic Digital Computer: Part 2-M. Nunn and C. G. Wynne. (Proc. Phys. Soc., vol. 74, pp. 316-329; September 1, 1959.) A description of the program used and the results achieved. Part 1-2483 of 1959 (Wynne).

CIRCUITS AND CIRCUIT ELEMENTS 621.3.049.7 409

Miniaturization and Micro-Miniaturization -G. W. A. Dummer. (Wireless World, vol. 65, pp. 545-549; December, 1959.) Description of new techniques giving increased reliability

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Abstracts and References

Compiled by the Radio Research Organization of the Department of Scientific and Industrial Research, London, England, and Published by Arrangement with that Department and the *Electronic* and Radio Engineer, incorporating Wireless Engineer, London, England

NOTE: The Institute of Radio Engineers does not have available copies of the publications mentioned in these pages, nor does it have reprints of the articles abstracted. Correspondence regarding these articles and requests for their procurement should be addressed to the individual publications, not to the IRE.

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ACOUSTICS AND AUDIO FREQUENCIES 534.2:621.374.5 367

Propagation of Sound in Plate-Shaped Solid Delay Lines—P. M. Sutton. (J. Acoust. Soc. Amer., vol. 31, pp. 34–43; January, 1959.) An analysis by diffraction theory. Quantitative results are compared with experimental data obtained from long plate-shaped fused-Si delay lines.

534.232-8:537.228.1 368 Electrical Equivalent Circuits and Vibration Patterns of Barium Titanate Transducers—R. Leisterer. (*Arch. elekt. Übertragung*, vol. 12, pp. 515-526 and 557-561; November and December, 1958.) Equivalent circuits of the elementary cube in the uniaxial and biaxial states of stress, derived from simplified piezoelectric equations, are used to obtain the characteristics and equivalent circuits for thin plate and tubular resonators. Experimental investigations on the latter are described, and optimum dimensions of circular disk resonators are given.

 534.232-8:537.228.1
 369

 Ultrasonic Barium Titanate Adhesion and

 Paste Transducers—A. Lutsch. (Nature, vol.

 184, pp. 1458-1460; November 7, 1959.)

 Methods of manufacture of two types of transducer are described and their characteristics are discussed.

A list of organizations which have available English translations of Russian journals in the electronics and allied fields appears at the end of the Abstracts and References section. The Index to the Abstracts and References published in the PROC. IRE from February, 1958 through January, 1959 is published by the PROC. IRE, May, 1959, Part II. It is also published by *Electronic and Radio Engineer*, incorporating *Wireless Engineer*, and included in the March, 1959 issue of that journal. Included with the Index is a selected list of journals scanned for abstracting with publishers' addresses.

 534.283-8:546.87
 370

 Ultrasonic Attenuation in Bismuth at Low

 Temperature—D. H. Reneker. (Phys. Rev.,

 vol. 115, pp. 303-313; July 15, 1959.) Experimental results and interpretation.

534.78 371 Intelligibility of Peak-Clipped Speech at High Noise Levels—I. Pollack and J. M. Pickett. (J. Acoust. Soc. Amer., vol. 31, pp. 14-16; January, 1959.) There is no loss of intelligibility when "clean" speech is peakclipped.

621.395.61 372 High-Quality Microphones—M. L. Gayford. (*Proc. IEE*, Part B, vol. 106, pp. 501–513; November, 1959. Discussion, pp. 513–516.) A review of modern practice in the design and operation of high-grade microphones.

621.395.616 373 The Uniformity of the Characteristics of a Standard Microphone—T. Hayasaka, M. Suzuki and I. Nakano. (*Rep. elect. Commun. Lab., Japan*, vol. 7, pp. 97–99; April, 1959.) Permissible tolerances in the manufacture of the Type MR-103 condenser microphone are related to deviations from the mean sensitivity curve.

621.395.616.089.6 374 Self-Reciprocity Condenser Microphone Calibration—T. Hayasaka, M. Suzuki and R. Araki. (*Rep. elect. Commun. Lab., Japan*, vol. 7, pp. 71–72; March, 1959.) A calibration method is described which uses a single microphone. Results are compared with those obtained with the three-microphone reciprocity method and show a difference within 0.2 db at frequencies below 6 kc and 1.5 db at 10 kc.

621.395.625.3 375 Investigation of the Recording Process in Magnetic Sound Recording: Part 1—G. Schwantke. (Frequenc, vol. 12, pp. 355–360; November, 1958.) A generalization of Westmijze's model (see, e.g., 1297 of 1954) is derived which is applicable to a wider range of hysteresis-loop shapes. Visual examination of magnetization curves obtained with special apparatus described shows the inadequacy of the theoretical assumptions.

621.395.625.3 On the Longitudinal Oscillations of Magnetic Recording Tapes—E. Belger and G. Heidorn. (Rundfunktech. Mitt., vol. 3, pp. 51-55; February, 1959.) The characteristics and causes of longitudinal tape oscillations are investigated using the noise sidebands of a sliding tone. Methods of reducing the oscillation are suggested.

ANTENNAS AND TRANSMISSION LINES 621.372 377

Diffraction of Surface Electromagnetic Waves at an Impedance Discontinuity—N. G. Trenev. (*Radiotekh. Elektron.*, vol. 3, pp. 27– 37; January, 1958.) An investigation of the diffraction of *E* and *H* waves at an impedance discontinuity. Expressions are obtained for the reflection and transmission coefficients. The diffraction field is determined and radiation characteristics are plotted. Diagrams show the radiation at the impedance discontinuity for different values of phase velocity.

621.372.2 378 Attenuation of Electromagnetic Waves Propagating along a Wire Helix Line—S. Kh. Kogan. (*Radiotekh. Elektron.*, vol. 4, pp. 181-186; February, 1959.) A characteristic equation is derived showing the dependence of the propagation constant on helix geometry and wavelength.

621.372.8 379 On the Theory of Tapered Waveguides— V. L. Pokrovskií, F. P. Ulinich and S. K.

V. L. Pokrovskii, F. P. Ulinich and S. K. Savvinykh. (*Radiotekh. Elektron.*, vol. 4, pp. 161–171; February, 1959.) Investigation of em wave propagation in a plane waveguide with finitely small degree of tapering. By combining the perturbation theory with the WKB method, it is possible to compute the reflection and scattering effects to any degree of approximation of a small parameter. Reflection and scattering depend to a large extent on the smoothness of the taper.

621.372.8:621.39:061.3380Long-DistanceTransmissionbyguide-(Proc. IEE, Part B, vol. 106, suppl.no. 13, pp. 1-199; 1959.)The text is given ofpapers presented at the IEE Convention heldin London, January 29 and 30, 1959.

621.372.83 381 Overmoded Waveguides—L. Solymar. (*Electronic Radio Eng.*, vol. 36, pp. 426–428; November, 1959.) Assuming the perturbation of the electric field negligible, a general expression is derived for the amplitudes of the spurious modes generated at the discontinuity between a circular waveguide supporting the H_{01} mode and a noncircular one. It is seen that no *E* modes are excited at the discontinuity.

621.372.83

Reflections at Waveguide Flange Joints-U. v. Kienlin and A. Kürzl. (Nachrichtentech.

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help of control theory. These mathematical laws are useful for understanding the operation of the system concerned, and especially valuable for predicting the existence of hitherto unknown structural and organizational relationships, which can be confirmed experimentally.

The electronic analog computer, being fed information about the respiration of the subject, in the form of an electric signal proportional to chest circumference, is able to calculate the timing of the heart beat from beat to beat. The correctness of the predicted heart rate is checked by recording the artificial heart rate simultaneously with the actual heart rate of the subject. The close correspondence of the predicted and actual changes of heart rate for a wide variety of modes of breathing, and for different individuals, proves the validity of the differential equations describing the phenomenon.

Individual variations, as well as the effects of age and of various drugs, are expressible in terms of the parameters of the equations.

Physiologically, the equations clarify the responsible nerve processes. The mathematical description of the reflex control of the heart rate through respiration gives for the first time a quantitative analysis of this phenomenon. which involves the central nervous system. Nerve impulses from at least two different stretch receptors within the chest are sent to the brain, in turn causing nerve impulses from the brain to the heart, along the vagus nerve, to be modified according to definite laws. These laws indicate as a consequence that it is stretch receptors and not haemodynamic factors which are responsible for initiating the changes in heart rate. It is shown further, mathematically, how the pacemaker of the heart responds to various degrees of inhibition produced by the action of the vagus nerve.

The analysis allows the heart rate effects of exercise and emotional stresses, through the action of the autonomic nervous system, to be more precisely perceived as clearly separated from the respiratory effects.

The respiratory effects are shown to be caused by two separate reflexes each producing biphasic heart rate transients in the same directions. The observed effects are the result of superposition of these transients. The variable and often paradoxical results previously observed in attempting to relate heart rate to respiration on a steady-state, nondynamic basis are thus explained.

A New Standardized and Calibrated Phonocardiographic System—A. A. Luisada and R. Zalter (p. 15)

A phonocardiographic system has been designed and built using standard units of known frequency characteristics.

The system is *calibrated*, as the relative and absolute intensity of the signal can be completed from the records' amplitude, given the conversion factor of the transducer and the degree of amplification in decibels. The system is *standardized*. It allows the spectrum to be scanned in bands of 1 octave of 3-db attenuation (therefore ideally flat) with the low-cut-off frequency coinciding with the nominal frequency.

A 36 db per octave attenuation for the highpass section is suggested for the selective recording of the high-frequency vibration of low intensity. Representative phonocardiographic tracings are presented.

Biological Flow and Process Tracing Using Nuclear and Electron Paramagnetic Resonance --J. R. Singer (p. 23)

A general discussion of the application of nuclear and electron paramagnetic resonance to determination of blood flow velocities in intact humans is discussed. The feasibility has been proven using mice. The system of measurement essentially utilizes paramagnetic (nonradioactive) tracers. One simple tracing scheme which has been employed experimentally is to saturate the protons in the blood stream and detect the density of these protons as variations occur due to flow. By this means the flow velocity is readily measured. A discussion of paramagnetic resonance theory precedes the discussion of the biological applications.

A Pulse Power Amplifier for Biological Stimulation—Harry Ludwig (p. 29)

A power amplifier designed for use as a biological stimulator is described. It is driven by a dc voltage or a pulse generator with an amplitude of 50 volts. The output power of up to 250 volts and 100 ma is precisely controlled. The rise time is less than 5 μ sec. A dual-channel version whose outputs may be independently controlled and mixed is also described. The circuit for the +400 volts and -300 volts regulated power supplies is shown.

The Use of Electronic Computers in Medical Data Processing—R. S. Ledley and L. B. Lusted (p. 31)

Some of the potential advantages of computer aids to medical data processing are: making available to the physician quantitative methods in areas relating to data analysis and differential diagnosis; assisting in the evaluation of the best alternative courses of action during stages of the diagnostic testing processes; making easily available to the physician reference to the most current information about new preventive measures, and diagnostic and therapeutic techniques; and periodic recording and evaluating of individual physiologic norms for the more sensitive determination of an individual's health trend relative to disease prevention.

A Unified Concept of Health and Disease— George L. Engel (p. 48)

Military Electronics

Vol. MIL-4, No. 1, January, 1960

Frontispiece and Guest Editorial- Rawson Bennett (pp. 1-2)

Spacious Fantasies—John R. Pierce (p. 3) Microwaves in the Space Age—II. Richard Johnson (p. 5)

Consideration is given to the direction of future microwaye device research and development based on requirements for space communication, tracking, search and surveillance, and interference and countermeasures. The crucial element in two-way communication with a space vehicle is the vehicle-borne transmitter. The large values of one-way transmission loss associated with interplanetary distances will require great improvements in these transmitters. Radar involves two-way transmission loss, so even more improvement will be required for tracking and search radars. Solution of interference problems will require increased instantaneous bandwidths. Accordingly, it is suggested that much additional microwave device research and development work is needed. Average power and efficiency of microwave tubes must be increased, with the longer centimeter wavelengths to be used for earth-based radars and 1-30 millimeter wavelengths to be used for radars to be borne in space vehicles. This work should concentrate mainly on traveling-wave tubes. Solid-state low-noise receiving devices and low-noise traveling-wave tubes must be extended to higher frequencies and broader bandwidths. Means of generating microwave power through the use of plasmas must be investigated. All types of microwave election devices must be adapted to the space environment.

The Role of Radar in Space Research— K. J. Craig, et al. (p. 11) A Radio-Astronomy Project at the University of Illinois-George C. McVittie (p. 14)

High-Altitude Measurements of X Rays and far Ultraviolet Radiation—Herbert Friedman p. (18)

Since 1946, the Naval Research Laboratory has conducted basic research in the physics of the upper atmosphere by means of high-altitude rockets. The program has emphasized all areas of research, including atmospheric structure and composition, the ionosphere, airglow and aurora, meteors, cosmic rays, and rocket astronomy. In the last area, which includes X ray and ultraviolet radiation measurements. NRL scientists have contributed a major portion of the experimental information available today. These comprise all the existing data on solar X rays, the X ray and ultraviolet emissions of solar flares, the first spectrogram of the sun covering the ultraviolet region below 3000 A and subsequent extensions of the spectrum into the extreme ultraviolet, the first quantitative measurements of solar Lyman- α , and the discovery of ultraviolet nebulosities and the Lyman- α glow of the night sky. A recent success in photographing the profile of Lyman- α with very high resolution opens the way to the use of optical resonance absorption as a gauge of atmospheric composition. This method may prove to be a most powerful technique for analysis of the very high atmosphere, well beyond the range of satellite drag measurements. The purpose of this paper is to describe the experimental approach used in accomplishing the radiation measurement program just outlined.

A Gas Cell "Atomic Clock" as a High-Stability Frequency Standard-Maurice Arditi (p. 25)

Precision Optical Tracking of Artificial Satellites—W. F. Hoffmann, et al. (p. 28)

Field Emission, A Newly Practical Electron Source—W. P. Dyke (p. 38)

The properties of the field emission electron source are discussed. These include high current density, small size, no heater, instantaneous response, and a highly non-linear currentvoltage relationship. Engineering data are then derived including conductance, perveance, beam power, etc. It is shown that the field emission cathode is electrically stable and that it has long life given suitable environments and/or operating conditions which are specifield. Microwave, voltage control and measurement, electron optical and other applications are discussed. A 300-megw flash X-ray tube now in production is described. The availability of the field emitter as a newly practical electron source is expected to make possible a number of new devices which may more often complement than compete with existing technology.

The ONR Program in the Electronics Research Laboratory of the University of California, Berkeley—Samuel Silver (p. 46)

Communication Using Earth Satellites— Jerome B. Wiesner (p. 51)

A review of the use of earth satellites for reliable, ionospheric-independent communication circuits includes considerations of losses in the propagating path, directivity features, and influences such as Doppler shift. The effects of such influences on bandwidth and range are illuminated.

Contributors (p. 58)

Correction

In the Abstracts of IRE Transactions of the December, 1959 issue of PROCEEDINGS OF THE IRE, the article, "The Nesistor—A Semiconductor Negative Resistance," was incorrectly attributed to Robert E. Nelson. The author of the paper, which appeared in the July, 1959 issue of ELECTRON DEVICES, is Robert G. Pohl. periments. Comparisons with cross modulation in amplifier tubes are made.

Experimental Notes and Techniques (p. 468)

Contributors (p. 470) Annual Index, 1959 (Follows p. 472)

Engineering Writing and Speech

Vol. EWS-2, No. 3, December, 1959

Editorial (p. 65)

Talking and Writing About Science-Sir Lawrence Bragg (p. 68)

Many of the pitfalls inherent in talking and writing about science can be avoided if one but heeds the admonitions presented here.

The Greeks Had a Word for It—Arnold P. G. Peterson (p. 72)

Numerical mileposts have changed down through the ages, always in the hope of providing a more universally understood and acceptable system. Here is another step in this direction.

Turning Engineers into Authors—Walter A. Murphy (p. 74)

Many companies have developed, or are in the process of developing, programs that seek to help engineers and scientists write up their work for publication. This article shows one approach to this problem.

Keys to Good Article Writing-John M. Carroll (p. 77)

Frequent and effective publication in engineering magazines have helped many engineers to fame and fortune. Here are a few suggestions to increase the chances of acceptance and the readership of your article.

A Review of the IRE-PGEWS-T. T. Patterson (p. 82)

How does PGEWS stand after two-and-onehalf years of operation? What future course should be taken? The present National Chairman expresses his views.

Some Legal Considerations in Presenting Technical Information—Robert II. Rines (p. 83)

Everyone who has ever written or presented a paper, or plans to do so, should have some idea of the legal problems involved. Here is some must reading!

Information Gaps and Traps in Engineering Papers—Herbert B. Michaelson (p. 88)

Information faults in manuscripts for professional journals are caused by two kinds of structural defects: the information gap, in which essential information is omitted, and the information trap, in which important points are not emphasized and are obscured in the paper.

Book Review—J. D. Chapline (p. 92) Contributors (p. 93)

Medical Electronics

Vol. ME-6, No. 4, December, 1959

DECEMBER, 1959

Introduction—Lee B. Lusted (p. 193) Foreword—F. L. Dunn and H. G. Beenken (p. 194)

An Ultrasonic Flowmeter—J. F. Herrick

and J. A. Anderson (p. 195) Design Considerations for Ultrasonic Flow-

meters—William R. Farrell (p. 198) Resumé of Discussion Group on Ultrasonic Flowmeters—J. F. Herrick (p. 202)'

A Pulsed Ultrasonic Flowmeter—D. L. Franklin, et al. (p. 204)

A pulsed ultrasonic flowmeter has been developed specifically for the simultaneous measurement of blood flow through various major blood vessels in the intact unanesthetized animal. The flow section is a small (-3 cm) lucite cylinder which is clamped about the blood vessel. Piezoelectric crystals are mounted on the flow section so that bursts of 3-mc sound may be transmitted alternately upstream and downstream. The flowmeter develops a voltage which is proportional to the difference in the upstream and downstream transit times of the sound. This voltage is recorded continuously and calibrated in terms of flow. Under optimal conditions, the output voltage is a linear and accurate representation of volume flow within ± 5 per cent, independent of the velocity profile. The flowmeter responds to a step variation in flow within 0.01 second. The maximum noise and baseline drift is equivalent to a flow velocity variation of less than 1 cm/second measured over a 4-hour period.

Comparative Pulsatile Blood Flow Contours Demonstrating the Importance of RC Output Circuit Design in Electromagnetic Blood Flowmeters—T. Cooper and A. W. Richardson (p. 207)

Measurement of Cardiac Output in Unrestrained Dogs by an Implanted Electromagnetic Meter—Frederick Olmsted (p. 210)

Gated Sine-Wave Electromagnetic Flowmeter—A. Westersten, et al. (p. 213)

Chopper-Operated Electromagnetic Flowmeter—Francis L. Abel (p. 216)

The problem of design of a square-wave electromagnetic flowmeter may be simplified by the use of a mechanical chopper as the gating device. Such an instrument has been assembled using commercial components costing less than \$1000. A commercial 400-cycle chopper amplifier, Offner model 190, was modified to perform the sampling. Carrier signal was provided by Tektronix 162 and 161 generators. Proper synchronization with the chopper is obtained by use of the 161 delay control. Magnet assemblies were made of ferrite toroidal cores and platinum electrodes, cast in epoxy resin and powered directly by a Heathkit 70watt amplifier.

Using a carrier frequency of 400 cps, good frequency response, stability, and signal-tonoise levels were attained. Calibration curves are linear for forward and backward flow. Sensitivity was sufficient to record blood flow levels of 1 ml per minute.

The Square-Wave Electromagnetic Flowmeter: Theory of Operation and Design of Magnetic Probes for Clinical and Experimental Application—M. P. Spencer and A. B. Denison, Jr. (p. 220)

Electromagnetic Blood Flow Measurements in Extracorporeal Circuits—A. R. Cordell and M. P. Spencer (p. 228)

A Magnetic Flowmeter for Recording Cardiac Output—H. W. Shirer, et al. (p. 232)

An Integrating Drop-Flowmeter for Optical or Pen Recording—C. N. Peiss and R. D. McCook (p. 234)

Performance and Application of a Commercial Blood Flowmeter—W. Thornton and B. Bejack (p. 237)

Design goals and reasons for the selection of a 60-cps square-wave electromagnetic blood flowmeter are presented. Detailed description of the flowmeter is given with emphasis on demodulator techniques. Complete performance data of step function response, stability, and performance with small and large flows are shown graphically. One application of the flowmeter and analog integrator is given showing response to a periodic hydraulic flow.

The DC Electromagnetic Flowmeter and Its Application to Blood Flow Measurement in Unopened Vessels—W. Feder and E. B. Bay (p. 240)

Flowmeter for Extracorporeal Circulation-G. Albertal, et al. (p. 246)

Discussion of Orifice-Plate Flowmeter for Extracorporeal Circuit—F. Robicsek (p. 249) Resumé of DC Electromagnetic Flowmeter Group Discussion—W. Feder (p. 250)

Certain Aspects of Hydrodynamics as Applied to the Living Cardiovascular System— Donald L. Fry (p. 252)

The Measurement of Pulsatile Blood Flow by the Computed Pressure Gradient Technique—Donald L. Fry (p. 259)

Methods of Flow Estimation by Pressure Sensing Techniques—Donald L. Fry¹ (p. 264) Nonimpromptu Comment on Papers by Dr.

Fry—Robert L. Evans (p. 266) Blood Flowmeter Utilizing Nuclear Mag-

netic Resonance—R. L. Bowman and V. Kudravcev (p. 267) The Potter Electroturbinometer : An Instru-

in the Dog—S. J. Sarnoff and E. Berglund (p. 270)

The flowmeter described is a turbine which can be driven by the bloodstream. In the turbine is a magnet which induces a recordable signal in an adjacent coil. The flowmeter records pulsating flow as well as steady flow; it is insensitive to temperature and wide variations of blood viscosity. The pressure drop is rather high. The base line and calibration are steady over long periods of time. The flowmeter has been successfully used for continuously recording the systemic output in the dog.

An Automatic Recording Bubble Flowmeter -C. W. Nash and J. V. Milligan (p. 274)

Use of Indicator Concentration Curves in Computation of Mean Rate of Flow and Volume of Blood Contained within a Segment of the Vascular System—II. D. Green, *et al.* (p. 277)

Isothermal Blood Flow Velocity Probe—S. Katsura, et al. (p. 283)

A New Velocity Probe for Sensing Pulsatile Blood Flow—A. MacDonell Richards (p. 286) Quantitative Measurement of Branched

Flow by Externally Placed Radioisotope Detectors—S. Thompson, *et al.* (p. 287)

Harmonic Analysis of Frequencies in Pulsatile Blood Flow-D. J. Ferguson and H. S. Wells (p. 291)

Critical Review of Bristle Flowmeter Techniques—Gerhard A. Brecher (p. 294) Index (Follows p. 304)

Medical Electronics

VOL. ME-7, NO. 1, JANUARY, 1960

Editorial-Lee B. Lusted (p. 1)

Respiratory Control of Heart Rate: Laws Derived from Analog Computer Simulation— Manfred Clynes (p. 2)

The application of dynamic systems analysis, essential for the design of controlled missiles and of man-made automatic control systems of all kinds, to the control systems within the human organism, is beginning to bear fruit in medical science. These organic control systems, which are part of the living process, generally far surpass any man-made systems in sublety and ingenuity. Their exploration in terms of feedback control theory is highly instructive both to the medical scientist and to the control engineer, since these systems generally display a quality and perfection of control design several orders of magnitude greater than the human mind can now conceive.

Some aspects of the control of the human heart rate are presented here. Normal and irregular respiration widely changes the rate of the heart from beat to beat. The mathematical laws describing this behavior were derived from analog computer simulation. On the analog computer a mathematical model of the dynamics of the processes involved was created. This was achieved through the use of dynamic data obtained from experiments designed by the

the high-current-density beam by the probegrid. This paper presents the design procedure and experimental results for typical probegridded guns. The design procedure is used to obtain the desired perveance, beam diameter, and approximate laminar electron flow. The probe geometry that results in a minimum beam distortion is discussed. The range of values of amplification factor obtainable and the influence of probe geometry on this factor are discussed. The magnetic field required for focusing the beam from a probe-gridded gun is compared with that required for perfect laminar flow and for focusing the beam from a nongridded gun of similar design. An electrolytic tank in conjunction with an analog computer was used to plot electron trajectories, with the effect of space charge included, for the probegridded gun and a similar nongridded gun. A comparison of the electron optics of the gridded and nongridded gun is made. Electrical breakdown and beam current during the interpulse time are problems considered. Methods used to minimize electrical breakdown and interpulse beam current are presented. Several models of probe-gridded guns were constructed. The measured characteristics of these guns demonstrate that the advantages of grid control can be obtained with only a minor effect on gun perveance and beam focusing.

The Effects of Initial Electron Velocities and Space Charge in Secondary Emission-M. D. Hare (p. 397)

This paper treats a secondary emitter as a fixed-temperature thermionic emitter with an equivalent work function which depends for its value upon the current density of the incident primary electrons. This permits Langmuir's treatment of the parallel-plane thermionic diode to be applied to secondary emission. The resulting equations account quantitatively for observed secondary-emission effects caused by space charge and initial electron velocities. The paper concludes with a discussion of three specific electron devices in which secondaryemission effects due to space-charge and initial electron velocity are important.

Space-Charge Layer Width in Diffused Junctions-R. M. Scarlett (p. 405)

This paper outlines a calculation of spacecharge layer width in a planar junction made by diffusing an n or p impurity (assumed to follow a Gaussian or a complementary error function distribution) into a uniformly doped crystal of opposite conductivity type. The collector junction of most drift transistors conforms closely to this model. An exponential approximation to the impurity distribution permits curves to be drawn of the space-charge layer penetration in each direction from the junction as a function of applied reverse voltage, and of the electric field distribution. The quantities involved are normalized in terms of the initial doping level N_1 , the impurity diffusion length $L = 2\sqrt{Dt}$, and the junction depth x_i . The curves should be useful in calculating depletion-layer capacitance, transistor punchthrough voltage and junction breakdown voltage.

Electron Beam Characteristics in Radially Varying Periodic Magnetic Fields-C. C. Johnson (p. 409)

Periodic magnetic fields are being widely used for the light-weight focusing of beams in high-power traveling-wave tubes. In many tube designs, there exists a considerable amount of radial variation in the magnetic focusing field. The effect of this radial field variation is investigated analytically as an extension of the previous work in this field. The usual design curves of α vs β are plotted with three variable parameters: ripple, cathode shielding parameter K, and radial field variation parameter x_0 . It is noted that it is important to keep the magnetic-field strength constant at the beam edge

over a considerable variation of the magneticfield parameter x0.

The Effect of Secondary and Backscattered Electrons in the Parallel-Plane Diode-L. A. Harris (p. 413)

Separate calculations are carried out to determine the influence of true, low-energy secondary electrons and of higher-energy backscattered electrons released from the anode of the parallel-plane, space-charge limited diode. Both groups depress the potential, increase the field near the anode, and decrease the net diode current by small but appreciable amounts.

Theory of the Amplitron-G. E. Dombrowski (p. 419)

The Amplitron device is analyzed from a normal mode viewpoint based on predominance of the reentrant character of the stream; the analysis is therefore valid for devices with short electron recirculation times and moderate signal levels. The nature of the space charge deduced from the above hypothesis is that of a rotating set of identical spokes having equal angular spacing in the interaction space.

The induction effects of this space charge configuration upon the delay line are calculated, accounting for the periodic nature and the short length of the delay line. It is found that both forward and backward-traveling waves are appreciable and must be considered.

The fields in the interaction space are resolved into Fourier component traveling waves. The amplitude of the synchronously interacting wave is related to 1) the input signal, 2) the forward-traveling wave resulting from space charge induction, and 3) the backward-traveling wave resulting from space charge induction. Use is made of the phase relation (adiabatic theory) between the space charge and the synchronously traveling wave to obtain a consistent solution determining the phase relation between the input wave and the space charge.

The above relationship between the space charge and the input signal allows the calculation of complex (vector) gain of the Amplitron. It is thus shown that the Amplitron is a nonlinear, or saturated, amplifier. A limit to the gain is observed; the backward-traveling wave is essential in determining it. Phasedependence on operating RF level, or RF phase pushing, is noted; this type of phase variation does not exceed 90°. Calculations as a function of frequency show the bandwidth to be expected. It is found that conditions may lead to oscillation; feedback mechanisms reside in the backward-traveling wave and in the stream reentrance. The degree of input mismatch of the operating tube is discussed.

Low-Noise Klystron Amplifiers-R. G. Rockwell (p. 428)

The principles of low-noise guns have been applied to klystron amplifiers with good corroboration of the theory. In the past, many people thought that klystrons had inherently high noise figures, while others advanced the theory that low-noise guns might be used with klystrons as well as with wave tubes. The development to be described here shows that the former impression is not true and verifies that low-noise klystron amplifiers are possible.

The most obvious difference between the guns for low-noise klystron amplifiers and those typical of low-noise traveling-wave tubes is the higher beam current which is required for adequate klystron gain. A byproduct of this higher current is a wide dynamic range.

In addition to the development of the electrical parameters, a major effort went into klystron construction techniques somewhat peculiar to low-noise klystron amplifiers. The data taken show that alignment of the lownoise gun electrodes with the drift tube, alignment of the beam with the magnetic field, elimination of the collector's secondary electrons from the beam, and cleanliness of the tube are

of primary importance in constructing a lownoise klystron amplifier.

Several two-cavity, low-noise klystron amplifiers were built for operation in both S-band and C-band. The typical low-level gain was 11.5 db, and the saturated power output was 180 mw. Several tubes exhibited noise figures below 9 db; the lowest value obtained was 6.7 db.

The Effect of Beam Cross-Sectional Velocity Variation on Backward-Wave-Oscillator Starting Current-N. C. Chang, et al. (p. 437)

Low-voltage helix-type backward-wave oscillators require a starting current that rises to infinity toward the low-frequency end of the tuning range. The effect has been attributed to the raising of the space-charge parameter OC by the dc space-charge-induced velocity spread. H. R. Johnson has calculated the velocityspread effect on starting current qualitatively, but the predicted non-oscillation frequency is generally much lower than the observed one. As a further analysis, space-charge wave propagation in an electron beam having an actual cross-sectional variation of dc velocity is investigated. It is shown that the RF current modulation in the slow spacecharge wave is concentrated in the region of the slowest-moving electrons. In a helix-type backward-wave oscillator using a hollow beam, the slower electrons are farther away from the RF circuit, so that the effective impedance for the slow space-charge wave may be considerably reduced. The use of an impedance reduction factor therefore provides better agreement between theory and experiment with regard to the starting-current phenomenon. Theoretical and experiment results of the investigation are presented.

The Cylindrical Field Effect Transistor-II. A. R. Wegener (p. 442)

The characteristics of a cylindrical fieldeffect transistor are derived analytically on the basis of Shockley's theory of the planar fieldeffect transistor. It is found that the cylindrical device is capable of giving twice the (voltage) amplification factor of that of the planar device. Its frequency behavior should be comparable to that of the Shocklev unit. Because of the loss of one degree of freedom, the transconductance and power characteristics of the cylindrical field-effect transistor are sharply limited. Experimental data support the analytical results.

Small-Signal Theory of Multicavity Klystrons-G. R. White (p. 449)

A small-signal formulation is developed which is valid for multicavity klystrons with nonideal gaps. The complete one-dimensional description of modulation at the gap is given, including the modifications due to space-charge forces. Stagger-tuned amplifiers are treated by matrix and by scalar methods. Equations useful for electronic computation of response are presented. The necessity for the formulation, and its validity, are discussed.

Cross Modulation and Nonlinear Distortion in RF Transistor Amplifiers-M. Akgun and M. J. O. Strutt (p. 457)

In order to avoid untractable calculations, the transistor four-pole is assumed to be shortcircuited for ac at its output. Furthermore, the internal impedance of the signal source is assumed to be zero. First the nonlinear distortion effects in a grounded base intrinsic transistor are calculated. Then, the formulas are reverted to a grounded emitter intrinsic transistor, taking into account the extrinsic base lead resistance. They are confirmed by measurements of third harmonic distortion and of cross modulation. The measured curves of cross modulation vs collector bias current show a sharp minimum. This unexpected effect is explained by an extension of the theory, which takes into account previously neglected terms. The explanation is successfully tested by ex-

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teristic depends only on two parameters. An explicit method is given by which the numerical values of the resistors used in the thyrite circuit can be computed for the given characteristics of the thyrite rod used in order to achieve desired "best" results.

Direct Single Frequency Synthesis from a Prescribed Scattering Matrix-D. C. Youla (p. 340)

This paper presents two techniques for synthesizing an *n*-port (at a single (requency) directly from its normalized scattering matrix without recourse to its associated impedance matrix. Both methods depend on standard matrix canonic forms. In the first one, all the loss is extracted immediately in the form of nresistors and the problem is reduced to the synthesis of a lossless 2n-port. The second makes use of the Jordan decomposition of a matrix and leads to a hybrid mixture of ideal transformers, reactances, gyrators and resistors. The only elements required for a complete synthesis are ideal transformers, lossless capacitors, inductors, and gyrators, and positive and negative resistors. It is shown that synthesis from a prescribed scattering matrix requires, in general, irrational operations (computation of eigenvectors, eigenvalues, etc.) whereas synthesis from a prescribed impedance matrix (if it exists) can be achieved with rational operations alone.

Minimal Realizations of the Biquadratic Minimum Function—Sundaram Seshu (p. 345)

The purpose of this paper is to obtain rigorously minimal realizations of the biquadratic minimum positive real function without the use of transformers. For this purpose a few theorems are proved about the structure of the network realizing a minimum *pr* function. This is followed by an exhaustive search of networks in increasing order of number of elements. It is proved that the modified Bott-Duffin (or the Reza-Pantell-Fialkow-Gerst) realization using seven elements is rigorously minimal in number of elements, except for the special cases $Z(0) = 4Z(\infty)$ and $Z(\infty) = 4Z(0)$. These two special cases have five element realizations.

On the Representation of Transients by Series of Orthogonal Functions—H. L. Armstrong (p. 351)

When problems having to do with transients are solved by the Laplace transform or equivalent methods, one may be left with the necessity of solving a rather complicated equation in the transform variable. This may be avoided, in many cases, by getting the solution in the form of a series. Laguerre functions have had some use for that purpose. It is shown here how another set of functions, which are just the lacobi polynomials whose argument is an exponential, may be used instead. The use of this latter set permits a rather elegant means of evaluating the coefficients in the expansion to be used. In an appendix, ways of applying the mathematical techniques used are investigated. These involve the complex "Faltung" theorem. for investigating questions of orthogonality and orthonormality in general.

Improving the Approximation to a Prescribed Time Response—John D. Brule (p. 355)

The topic considered in this paper is the problem of obtaining the Laplace transform of a prescribed impulse response, under the constraint that this transform must be a realizable rational function. In general, the solution of this problem requires that approximations be made, either in the time domain or in the frequency domain. A procedure is developed which provides a systematic method for improving the approximation by making small changes in the poles and residues of the transfer function. The effects of such changes on the impulse response are evaluated by means of a Taylor series expansion of the impulse response. It is shown that only the first two terms of this expansion provide a reasonably accurate estimate of these effects. A set of normalized curves are prepared which allow the designer to determine how a given pole or residue should be changed in order to improve the approximation in the time domain. The procedure is demonstrated by applying it to a numerical example.

An Extension of Wiener Filter Theory to Partly Sampled Systems--H. M. Robbins (p. 362)

The growing use of digital computers as components of control systems has given great importance to the study of linear systems which are partly sampled and partly continuous. This paper treats the problem of optimizing the simplest possible mixed system consisting of an input filter with transfer function K(s), a sampler with sampling interval T, and an output filter with transfer function L(s). Given the power spectra of the input signal and the noise, the object is to find a realizable K and L which in combination minimize the mean square difference between the output h and a "desired output" h_a , h_a is defined by a "desired transfer function" $G_d(s)$, not necessarily realizable, which would produce h_d from the input signal if the noise were absent.

KL will in general contain factors periodic in s with period $2\pi j/T$, and such factors may be moved to either side of the sampler without changing the final output, thus introducing a considerable arbitrariness in K and L. However, since these periodic factors represent linear operations on discrete data (such as might be performed inside a digital computer), it is appropriate to separate them out. There are then three functions to be determined: the nonperiodic part of K, the nonperiodic part of L, and the remaining (periodic) factor of KL. Methods for determining these three functions are given. The interesting theoretical point is that the determination is not always unique. In general, there will be a finite number of distinct but equivalent solutions.

Network Functions with a Constant Imaginary Part—H. J. Orchard (p. 370)

The paper considers the problem, suggested by the circuitry of the so-called seven-league oscillator, of constructing a positive-real rational function whose imaginary part makes an equal-ripple approximation to a constant value over a wide band of frequencies. It gives an exact solution using elliptic functions and shows that the resulting function satisfies the conditions for realization with either an RL or an RC circuit. Performance curves and computing details are included.

Generalizations of the Concept of the Positive Real Function—A. H. Zemanian (p. 374)

Two generalizations of the concept of the positive real function are made that are applicable to transfer functions whose poles outnumber their zeros by any amount. Some previously published results are briefly discussed. Then, the second generalization is developed still further. In particular, the properties of the zeros on the imaginary axis are established and are found to be analogous to the corresponding properties for positive real functions. In addition, several new tests for the generalized functions are developed, one of which serves as a new test for positive real functions wherein the Sturm sequence is replaced by a more easily calculated sequence. Reza's "double alternation" property for positive real functions is extended to the generalized functions. Finally, a property of the phase functions of the positive real functions is extended and the use of some common transformations is discussed which, in turn, leads to still another type of generalization.

Reviews of Current Literature (p. 383) Correspondence (p. 386)

The 1960 Solid-States Circuits Conference (p. 399)

Annual Index 1959 (follows p. 402)

Electron Devices

Vol. ED-6, No. 4, October, 1959

Large Signal Bunching of Electron Beams by Standing-Wave and Traveling-Wave Systems—S. E. Webber (p. 365)

Large signal bunching processes in the presence of space charge are studied by extending techniques used to compute multi-cavity klystron bunching. Bunching capability of traveling-wave and standing-wave systems is examined by using an assumed spatial and time distribution of electric field, and interesting high degrees of bunching are predicted.

A Theoretical Study of Ion Plasma Oscillations—W. W. Peterson and H. Puthoff (p. 373)

A theoretical study is made of oscillations in an ion plasma, which is in an electron beam. The effect of ion motion on the electrons is neglected, and the ions and the electrons are assumed to have constant and equal density in the equilibrium position. Symmetric and transverse oscillations are studied, both in planar and cylindrical geometry. For planar geometry, the frequency of oscillations for both symmetric and transverse modes is independent of amplitude, while the frequency increases with amplitude for cylindrical symmetric oscillations. For both cylindrical and planar geometry, the presence of the anode boundaries reduces the frequency for transverse ion oscillations. but does not affect the frequency for symmetric-type oscillations.

A New Electronic Gun for Picture Display with Low Drive Signals—Kurt Schlesinger (p. 377)

The paper describes an approach to the problem of picture tube guns for small signal service.

A Pierce-type cathode delivers a collimated parallel beam of 1600 microamperes at 250 volts (microperveance: 0.4). This beam is injected into a cylindrical cavity of appreciable length $(\frac{3}{4}$ -inch long, $\frac{1}{4}$ -inch diameter). It is focused upon a small aperture at the far end using a parabolic axis potential. This axial focusing field is approximated by three cylinder segments at two intermediate voltages.

To modulate the beam by lateral deflection, the cavity is again bisected through an axial plane and signal voltage is set up between half cylinders. This modulation by two crossedelectrostatic fields ("CFM" modulation) has been successfully applied in transistorized television, using a seven-volt video signal on a beam of 900 microamperes.

Diffused Silicon Nonlinear Capacitors-A. E. Bakanowski, et al. (p. 384)

Diffused silicon nonlinear capacitors have been fabricated by solid-state diffusion. The resulting graded p-n junction is a planar structure which permits low series resistance R_s relative to the minimum capacitance C_{\min} , which is measured at a reverse voltage slightly less than the breakdown voltage. A cutoff frequency $f_c = (2\pi R_s C_{\min})^{-1}$ is used as a figure of merit; values up to 150 kmc have been obtained.

These "varactor" diodes may be used as UHF and microwave amplifiers and as harmonic generators. The noise figures of the UHF amplifiers are better than the best noise figures obtainable by present electron-tube techniques. These diodes are also efficient harmonic generators.

The Design and Performance of Grid-Controlled, High-Perveance Electron Guns-H. E. Gallagher (p. 390)

The focusing electrode and a probe projecting through the cathode serve as control electrodes for the current from a convergent-beam electron gun. The principal advantage of this type of "grid" is that there is no interception of

mental of physiological measurements. Investigators have worked for more than three centuries to develop methods for measuring various parameters of the circulation of the blood. Despite this, there is no generally accepted method today for measuring volume of flow. Rather, a number of methods are available, all of limited applicability. In some instances various types of ingenious velocity or pressure sensing devices (in one case, a tiny blood-driven turbogenerator) are inserted directly into the blood vessel. Recent developments in the electronic art, however, have focused considerable attention on indirect methods which do not require opening the circulatory system. One technique involves producing and then detecting nuclear magnetic resonance of hydrogen protons in the blood. There is also much interest in flowmeters of the electromagnetic type. Here the blood, because it is an electrical conducting fluid, acts as a moving conductor which, in the presence of an electromagnetic field, will induce an EMF proportional to the velocity of flow. This same principle, incidentally, was used as far back as 1832 by Faraday to measure tidal changes in the Thames River. Another very promising method makes use of the fact that ultrasonic waves travel faster downstream than upstream by an amount related to the velocity of the blood. These and other techniques were the subject of an important conference held last June under the joint sponsorship of the University of Nebraska School of Medicine and the IRE Professional Group on Medical Electronics. The 26 pages presented at the conference have now been published by the PGME. Since the work of virtually all the leading investigators in this field is reported, the issue stands as an especially valuable reference to the bio-medical electronics community. (Blood Flowmeters Symposium Issue IRE TRANS. ON

MEDICAL ELECTRONICS, December, 1959.)

A new electron gun for TV picture tubes offers the possibility of the first major change in gun structure for television displays in many years. The gun has several unique design features, one of them being that the beam emanating from the cathode is unmodulated. Modulation is accomplished by lateral deflection of the beam, causing it to be partially intercepted in a controlled manner as it passes through an aperture. The most important difference between this and standared guns, however, is that it will operate with a much lower video signal—about 7 volts—and thus is eminently suited for use in transistorized sets. (K. Schlesinger, "A new electron gun for picture displays with low drive signals." IRE TRANS. ON ELECTRON DEVICES, October, 1959.)

Klystron amplifiers have now joined the low-noise club. Many people have considered klystrons to be inherently noisy Others have believed otherwise, theorizing that noise limitations were basically no different for klystrons than for conventional traveling-wave and backward-wave tubes. However, there was no experimental proof of the theory. Meanwhile, klystrons continued to be noisy. The theorists, who maintained that the noise figure is determined mostly by the gun and not by the microwave circuit, have now been vindicated. The principles of low-noise guns, developed earlier for wave tubes, have been applied successfully to klystron amplifiers to produce tubes with noise figures as low as 6.7 db. This experimental verification opens the door to the use of tuned radio-frequency receivers for radar or radio astronomy, having the combination of high sensitivity, high gain, and wide dynamic range. (R. G. Brockwell, "Low-noise klystron amplifiers," IRE TRANS. ON ELECTRON DEVICES, October, 1959.)

Abstracts of IRE Transactions_

The following issues of TRANSACTIONS have recently been published, and are now available from the Institute of Radio Engineers, Inc., 1 East 79th Street, New York 21, N. Y. at the following prices. The contents of each issue and, where available, abstracts of technical papers are given below.

Sponsoring Group	Publication	Group Members	IRE Members	Non- Members*
Circuit Theory	CT-6, No. 4	\$1.35	\$2.00	\$4.05
Electron Devices	ED-6, No. 4	2.20	3.30	6.60
Engineering Writing	-,			0.00
and Speech	EWS-2, No. 3	1.15	1.70	3.45
Medical Electronics	ME-6, No. 4	2.40	3.60	7.20
Medical Electronics	ME-7, No. 1	1.50	2.25	4.50
Military Electronics	MIL-4, No. 1	1.15	1.70	3.45

* Libraries and colleges may purchase copies at IRE Member rates.

Circuit Theory

Vol. CT-6, No. 4, December, 1959

Analysis of Periodic Filters with Stationary Random Inputs—Harry Urkowitz (p. 330) A periodic filter is one whose frequency characteristic is periodic in frequency. Examples are delay-line cancellers for moving-target indication and delay-line sweep integrators. Ztransform techniques have proved useful in the analysis and synthesis of these filters when determinate inputs are involved. The determination of the mean square output and other

statistical properties with random inputs has heretofore usually involved numerical integration, even for fairly simple cases. This paper presents formulas for computing the mean square value and other statistical properties of the output when the input is a stationary random function. The formulas involve the values of the input autocorrelation function at integral multiples of the basic delay and the sums of product pairs of the coefficients in the Z-transform power series. Simple algebraic forms result, making slide-rule computation feasible. The effects of internal noise are also considered by means of the same techniques. Specific formulas are derived for single and double cancellers, velocity-shaped cancellers, and sweep integrators.

Generation of Squares with the Use of Non-linear Resistors—E. Grosswald $(p,\,334)$

Whenever great accuracy is not required, a convenient way to obtain squares, and hence, to multiply, in analog computers is by the use of nonlinear resistors, especially thyrite rods, in certain circuits.

These circuits are usually determined, more or less empirically, by starting from schemes that have given good results in the past and modifying them, or by varying the numerical values of the resistors used, until "best" results are obtained. In the present paper the concept of "best" results is clarified. Furthermore, it is shown that, regardless of the complexity of the circuit, the input-output charac-

Recent Books

- Bershader, Daniel, *The Magnetodynamics of Conducting Fluids*. Stanford University Press, Stanford, Calif. \$4.50. A review of the field of behavior of conducting fluids in magnetic fields.
- Besserer, C. W. and Hazel, *Guide to the Space Age.* Prentice-Hall, Inc., Englewood Cliffs, N. J. \$7.95. Presents in clear and simple language the terminology of space technology and guided missilery.
- Contini, Renato and Paul T. Bryant, Eds., Engineering College Research Review 1959. Engineering College Research Council of the American Society for Engineering Education, Urbana, Ill. \$2.00. A list of all research projects and research personnel at institutions holding membership in the Engineering College Research Council.
- Jastrzekski, Żbigniew D., Nature and Properties of Engineering Materials. John Wiley and Sons, Inc., 440 Fourth Ave., N. Y. 16, N. Y. \$11.00. Provides the reader with the basic knowledge necessary for intelligent selection and use of materials for special engineering applications and prepares him to solve materials problems of the future.

Lettenmeyer, Lore, Dictionary of Atomic

Terminology. Philosophical Library, Inc., 15 E. 40 St., N. Y. 16, N. Y. \$6.00. Provides a selection of the essential scientific and technical terms employed in connection with atomic and nuclear physics, reactor engineering, radiation physics and associated fields, with the objective of facilitating the study of the relevant foreign literature on the subject.

- Leinwoll, Stanley, *Shortwave Propagation*. John F. Rider Publisher, Inc., 116 W. 14 St., N. Y. 11, N. Y. \$3.90. Presents the basic principles of shortwave radio propagation, and how it is used in long distance radio communication.
- Nayler, J. L., Dictionary of Aeronautical Engineering. Philosophical Library, Inc., 15 E. 40 St., N. Y. 16, N. Y. \$10,00. An illustrated dictionary which provides concise definitions of aeronautical engineering terms.
- Proceedings of an International Symposium on the Theory of Switching (Parts I and II). Harvard University Press, Cambridge, Mass. \$15.00. Consists of 39 papers which were presented at the symposium, which was held at the Computation Laboratory of Harvard University, April 2-5, 1957.

Proceedings of the Fourth Midwest . Sym-

Scanning the Transactions_

Dilocycles and trilliseconds? An IRE Technical Committee once jokingly defined a kilocycle as a two-wheeled vehicle for transporting ten centipedes. While there can be no doubt in the mind of any electronics engineer as to the real meaning of kilocycle, the same certainly cannot be said of other units of measure which are in use today. Does everyone remember that picofarad means 10⁻¹² farad? Do you recognize nanosecond (10⁻⁹ second) or gigacycle (10⁹ cycle)? The problem of finding suitable numeric prefixes for very large and very small quantities is becoming increasingly vexing because the flow of technical developments is carrying us to new orders of magnitude which have not yet been adequately staked out in our language. We do not even have agreement on such common terms as billion, trillion and quadrillion. To the British they mean the second, third and fourth powers of a million (a much more logical system than the American system). Thus the British billion is the equivalent of the American trillion. We need a system that will give us numeric prefixes which are short, recognizable and logical over a broad range of values. If we examine what we have now in the range 10⁻⁶ to 10⁶ we find that Greek prefixes are used for multiples greater than one and Latin is used for submultiples, except micro. This leads very logically to the suggestion that we always use Greek for positive powers of ten and Latin for negative powers. The result is the interesting system shown below.

Value	Prefix	Symbol	Value	Prefix	Symbol
$ \begin{array}{c} 10^{3} \\ 10^{6} \\ 10^{9} \\ 10^{12} \\ 10^{15} \\ 10^{21} \\ 10^{21} \\ 10^{27} \\ 10^{30} \end{array} $	kilo dilo trilo pentilo hextilo heptilo oktilo enneilo dekilo	K D TR TT PN HX HP OK EN DK	$ \begin{array}{c} 10^{-3} \\ 10^{-6} \\ 10^{-9} \\ 10^{-12} \\ 10^{-15} \\ 10^{-18} \\ 10^{-21} \\ 10^{-24} \\ 10^{-27} \\ 10^{-30} \end{array} $	milli- billi- trilli- quadrilli- quintilli- septilli- octilli- nonilli decilli	m b qd qn sx sp oc nn dc

posium on Circuit Theory. \$7.00. Purchase from: Raymond Kipp, Marquette University, 1515 W. Wisconsin Ave., Milwaukee 3, Wis.

- Schure, A., R-F Amplifiers. John F. Rider Publisher, Inc., 116 W. 14 St., N. Y. 11, N. Y. \$2.40. Presents a discussion of amplifiers—low powered as well as high powered—used in the RF portion of the radio frequency spectrum. Special emphasis is placed on the properties of resonant circuits as applied to power amplifiers.
- Smith, Woodrow, Radiotelephone License Manual, 2nd ed. Editors and Engineers, Ltd., Summerland, Calif. \$5.00. This question and answer manual is intended to be used as an aid in preparing for examinations for the various grades of radiotelephone license or permit.
- Van Valkenburgh, Nooger and Neville, Inc., Basic Electronics—Vol. 6. John F. Rider Publisher, Inc., 116 W. 14 St., N. Y. 11, N. Y. \$2.90 (paperbound) \$3.60 (cloth binding). Companion volume to the present five-volume course on basic electronics; intended to enable individuals, schools, and industrial programs to expand into areas of semiconductors, transistors and frequency modulation.

(A. P. G. Peterson, "The Greeks had a word for it," IRE TRANS. ON ENGINEERING WRITING AND SPEECH, December, 1959.)

Radar astronomy, although intimately associated with radio astronomy, is rapidly assuming a role of sufficient importance to justify the distinction of a name of its own. It has its beginnings as far back as 1925 when the Carnegie Institution of Washington and the Naval Research Laboratory began probing the ionosphere by means of pulsed radio signals. Radar astronomy made its first leap beyond the earth's environs in 1946 when the Army Signal Corps contacted the moon. Another major breakthrough came in 1957 when M.I.T. Lincoln Laboratory reached Venus by radar. Only last month Stanford announced that it had contacted the Sun. These experiments have given us new information on interplanetary distances. Future radar soundings of Venus and certain asteroids will give us a very accurate check on the size of our solar system. However, the return signal of a radar can provide us with a good deal more information than simply distance. By carefully analyzing the return pulse, deductions can be made concerning such matters as the density and kinetic temperature of ionized gases, motions of electrons, the magnitude and direction of any magnetic field present, and the nature and composition of the surfaces of solid bodies. This gives us the wherewithal to learn much more about the sun, the moon, the planets and the space between. For example, we should in time be able to determine the height of mountains on the moon, the rotational period of Venus (which is obscured to us by clouds), and the velocity and course of the highly important streams of charged particles known to be flowing from the sun toward the earth. Recent progress in large antennas, powerful transmitters, very low noise receivers, and special signal processing techniques all point to an increasingly important role for radar in future space research, (K. J. Craig, et al., "The role of radar in space research," IRE TRANS. ON MILITARY ELECTRONICS, January, 1960).

The measurement of blood flow is one of the most funda-

The editors have indeed succeeded in creating an important reference work which summarizes far more subjects between two covers than are found in any other single publication on computers. This fact alone should secure a firm place for this volume next to the more specialized literature on the computer book shelf.

WERNER BUCHHOLZ IBM Corp. Poughkeepsie, N. Y.

Principles of Analog Computation, by G. W. Smith and R. C. Wood

Published (1959) by McGraw-Hill Book Co., Inc., 330 W 42 St., N. Y. 36, N. Y. 230 pages +4 index pages +viii pages. Illus. 61 ×91, \$7.50.

This book appears to be written at a senior undergraduate or first-year graduate engineering level. At this level, it should be a useful text.

The presentation of the material of the book is from the point of view of those using operational amplifier analogue computers for the solution of various physical problems; only enough material concerning design principles of the component apparatus of such computers has been given to make it plausible that the components should operate as stated.

The authors discuss in some detail problem programming; the simulation of linear systems; the use of diodes, functional relays, and various types of function generators; and the use of implicit function techniques. Included is a very good collection of sets of problems to be solved.

This should be a useful text at the level for which it was written. In light of the purpose for which it appears to be intended, this reviewer regrets to note that very little discussion has been given to the problem of obtaining suitable scale factors, as this appears to be one of the major areas in which those just learning the use of analogue computers tend to have considerable difficulty.

Tables 3-1 and 3-2 of the text, indicating, respectively, the manner in which computer apparatus may be interconnected for the realization of various transfer functions, and the transfer functions of various passive networks, should be of considerable utility even to those with more experience in the field than the beginning student; and the discussion of five-impedance networks for use with a single amplifier together with the associated nomographs should also serve as a convenient reference.

There is, perhaps, a little too much detail regarding matters more relevant to servomechanism theory than to analogue computer use.

The discussion of implicit function generation is well-presented and complete enough to permit application of the principles given to more complex cases than those considered in the text.

This text should form a valuable addition to the literature on analogue computers, particularly for those whose interests center on servo-mechanisms theory, and who are just beginning their studies in the field.

KENNETH B. TUTTLE Beckman Instruments, Inc. Los Angeles, Calif.

Analog Methods, 2nd edition, by Walter J. Karplus and Walter W. Soroka

Published (1959) by McGraw-Hill Book Co., Inc., 330 W 42 St., N. Y. 36, N. Y. 429 pages +7 index pages +44 appendix pages +xiii pages. Illus. 6½ X9½. \$12.50.

In recent years, stress on applications of digital computers has, to a certain extent, obscured the quiet and orderly parallel progress made on analog methods. One may easily lose sight of the possibility that cases exist where an analog computer may actually be superior to a digital one, and that there are other situations where the two methods may be advantageously used to complement each other. A book which gives an up-to-date description of the analog computer art is therefore a valuable addition to the library of anyone who needs computing aids. The present book fulfills this specification very well. The coverage is encyclopedic in nature. A complete description is given of almost every analog method which has ever been used, including both mechanical and electrical models with either lumped or distributed parameters. Copious references are furnished with each chapter. The development proceeds from the basic elementary operations of addition, subtraction, multiplication and division, through integration and differentiation, function generation, and solutions of algebraic and differential equations. Both ordinary and partial differential equations are treated, and for the latter. finite difference approximations are discussed as well as continuous models using stretched membranes, soap films, electrolytic tanks and conducting sheets.

The wide range necessitates some restriction in depth of penetration for individual topics. Matters such as stability and bandwidth limitations are treated in a rather rudimentary manner, making little use of the more sophisticated concepts of modern electric circuit theory. The tip-off in this respect is that there does not appear to be any mention of the names Nyquist and Bode. The discussion of organization of an analog computer program points out the advances that have been made in the use of problem boards to expedite rapid access to and relinquishment of the machine, as well as to preserve programs intact for subsequent use. After citing the advantages, the author comes up with this somewhat nullifying statement: "Even so, the programming of reasonably complex problems results in a veritable maze of plug-in wires, a factor which not infrequently leads to errors in programming and makes problem checking very difficult." If this is the case, it is not clear why such a condition should be allowed to persist. Surely, with widespread present day usage of such organizational aids as block diagrams and flow graphs, it should be possible to construct special purpose plug-in units for commonly used operational functions and thereby make the problem board layout readily identifiable with the graphical model.

The book appears to be well suited for textbook use at the upper undergraduate level. Illustrative problems and laboratory exercises are included.

> W. R. BENNETT Bell Telephone Labs. Murray Hill, N. J.

Principles of Optics, by M. Born and E. Wolf

Published (1959) by Pergamon Press, 122 E. 55 St., N. Y. 22, N. Y. 715 pages+28 index pages+58 appendix pages+xxvi pages. Illus. 61 ×10. \$17.50.

This book, although based in part on Born's Optik, must be considered a new book as it contains less than 50 per cent of the old material and 75 per cent of its contents are different from Optik. In fact, half of this book is basic electromagnetic theory and must be read and reviewed in that context. The first two chapters describe the properties of the electromagnetic field and the treatment is similar to the standard references. The third chapter describes the foundation of geometric optics. Chapter IV describes optical imaging and works into the geometric theory of aberrations (Chapter V), and into the basic principles of image forming instruments (Chapter VI), contributed by P. A. Wayman. Chapter VII describes the principles involved in the theory of interference and interferometers and was contributed by W. L. Wilcock. Chapter VIII introduces the diffraction theory which is used in Chapter 1X to derive the theory of aberrations from the diffraction theory viewpoint. This is followed by a particularly excellent chapter describing the theory and application of partial coherence (Chapter X). Chapter XI is a shortened treatment of exact solutions in diffraction theory contributed by P. C. Clemmow. Chapter XII, written by A. B. Bhatia, describes the diffraction of light by ultrasonic waves. Chapter XIII is based on the optics of metals and is followed by an excellent treatment in Chapter X1V of the optics of crystals. These last two chapters are revised versions of part of *Optik*. The revision and extension of these chapters was performed by A. M. Taylor and A. R. Stokes. The appendixes are excellent contributions in themselves. The first is on the calculus of variations and is based on D. Hilbert's unpublished notes. The second is on the analogy of light optics, electron optics and wave mechanics by D. Gabor. The third is by P. C. Clemmow on the asymptotic approximations to integrals. The fourth is on the Dirac delta function. The last five also elucidate further on mathematics in the text.

This book is excellently written, especially when one considers the variety of subjects and authors. It has been needed for many years. The major complaint must also be that it has been written over many years and thus, although some sections contain very new material, other sections end with the material as it existed five years ago.

The book will materially aid those workers who desire an over-all look into the mathmatical theory of optics. It can serve as a textbook for several courses and a reference book for others. It is recommended as an outstanding addition to the electromagnetic theory literature.

> KEEVE M. SIEGEL Univ. of Michigan Ann Arbor, Mich.

Books_

Principles of Electronics, by M. R. Gavin and J. E. Houldin

Princeton, N. J. 337 pages +6 index pages +1 bibliography page +5 appendix pages +xii pages. Illus. 5½×83. \$5.75.

This book is an excellent middle ground fundamental text on electronics. Neither highly theoretical nor very practical, it is clearly intended as a student text serving well both major groups of electronics students, *i.e.*, those who wish to have merely a general ground in the field and those desiring a good basic course prior to advanced specialization.

While Laplace's and Poisson's equations are given along with a number of differential equations, a very excellent understanding of the text can be achieved with no mathematical training beyond calculus. In fact, the text is sufficiently nonmathematical so that a "practical" engineer or technician with mathematics through algebra and trigonometry can get 75 per cent or more of its value.

The use of British standard symbols may lead to some confusion on the part of American students and engineers and it seems very doubtful to this reviewer that the use of this text in American universities will be popular; however, this is a very minor drawback for an American electrical engineer who wishes to get a good fundamental knowledge of electronics which he missed in college.

From an American standpoint it seems that considerably more material on solidstate electronics, particularly the transistor, would have been justified. However, to do the subject full justice at this level would require a major increase in volume, presumably not desired by the authors.

All in all it is a very excellent volume.

CONAN A. PRIEST 314 Hurlburt Rd. Syracuse 3, N. Y.

Fluctuation Phenomena in Semi-Conductors, by A. van der Ziel

Published (1959) by Academic Press, Inc., 111 Fifth Ave., N. Y. 3, N. Y. 164 pages +4 index pages +viii pages, Illus, $5\frac{3}{4} \times 8\frac{3}{4}$, \$6.50.

This book, as its title says, discusses noise in semiconductors. The book classifies noise into generation-recombination diffusion noise, and modulation noise, and proceeds to discuss examples of each kind.

The organization of the book, as far as the phenomenological discussion is concerned, proceeds from the simple to the difficult, with first a discussion of noise in semiconductor filaments, photoconductors, p-njunctions, and, finally, junction transistors, including a section on flicker noise.

In addition to the discussion of noise from a phenomenological viewpoint, there is a short discussion on each of the topics of noise measurements, circuit characterization of noise, general noise theorems, and circuit design for minimum noise.

The book is intended for students entering the field of noise study or workers in the field. Its chief virtue is the collection of a great deal of different but related material about noise in a single book, arranged by an experienced contributor to the field.

The book has a good, logical arrangement and includes accounts of the most recent advances in theories of noise. Perhaps the nost serious obstacle to easy reading lies in the completeness of the coverage of existing theory. The book would have more general usefulness if Professor van der Ziel had selected what he feels is the theory which is most likely to endure the test of time and had spent more time expounding his choice of the most favored theory.

DR. JOHN L. MOLL Stanford Elec. Labs. Stanford University Stanford, Calif.

Handbook of Automation, Computation and Control, Vol. 2, E. M. Grabbe, S. Ramo and D. E. Wooldridge, Eds.

Published (1959) by John Wiley and Sons, Inc., 440 Fourth Ave., N. Y. 16, N. Y. 1033 pages +37 index pages +xxiii pages. Illus. $6^+_8 \times 9^+_4$. \$17.50.

This handbook volume, the second of a series of three covering the field of computers and automatic control, deals with the design and application of digital and analog computers. With the help of an impressive list of 40 authors, the editors compiled 31 chapters covering the full range of subjects, from programming to design, from installation problems to inventory control, and from differential analyzers to game-playing machines.

One mammoth chapter alone, occupying fully one quarter of the book, treats the subject of programming digital computers. It covers the whole gamut of subjects which concern the programmer and ends with an extensive bibliography. The next nine chapters deal with digital computers as a whole and their application to business, scientific, and other areas. There follow nine more chapters on the design of components, circuits, and the major sections of digital computers. Another eight chapters on all aspects of analog computers occupy one sixth of the pages of the volume. Three more chapters deal with miscellaneous topics, and there is one chapter on terminology. Coverage is restricted to U. S. equipment and techniques, with the exception of one Soviet computer and one Soviet algebraic language compiler.

To publish a handbook on any broad subject is a tremendous undertaking, and it is not made any easier when the subject matter is continually undergoing major technological changes. The editors have chosen the basic structure well, allocating space to each area in fair proportion. They have not been quite so successful in guiding individual authors. Some chapters are truly excellent summaries of their field. Some other chapters are fairly uninspired essays containing more editorial comment and less reporting of facts than one would like to see in this book. Some amble past all the landmarks in their fields without conveying much information.

A few of the authors may have misunderstood the nature of a handbook. If they looked in Webster's dictionary, his definition of a handbook as "a book of reference to be carried in the hand" may have seemed inadequate when the average two-handed person is confronted by three thick volumes. Instead, paraphrasing the preface of another well-known handbook (Kent's) from the same publisher, one might better define a handbook as a book containing selections from an accumulated mass of data, condensed, digested, and arranged in a handy form for publication. There was no mention of opinions.

Overlap is unavoidable when a field is divided among many authors. One would expect to find certain basic material summarized in more than one chapter, but in several instances none of the two or three scattered synopses contain the real meat. Equally disconcerting is an occasional tendency to describe the past or to anticipate the future, at the expense of the proven techniques in widespread use. Thus, there are pages of detailed programming examples from an early computer which, having served its purpose, was dismantled two years ago; in contrast, only a few lines of summary data dispense with the most common type of large business computers, more of which are still being installed every year. Again, three pages are devoted to a fair treatment of rather tentative techniques for character sensing, but the omnipresent keypunch is dismissed with a quarter page. Tedious as it may seem to the erudite author, the bulk of a handbook should be devoted to mundane matters, however "old-hat."

One might also wish that some authors had taken a broader point of view. The editors obviously designed the structure of the book to reflect both of today's major computer applications: business data processing and mathematical computation. Individual chapters, however, do not strike a proper balance. The lengthy programming chapter virtually ignores the former category, except perhaps for a section on sorting. Conversely, eight of the nine chapters describing equipment, installation planning, and applications by-pass the scientific and engineering field. Fortunately, much valuable material, to be found in this volume and its predecessor, is basic to both of these and to other fields.

The section on analog computers is possibly the best example in this volume of what one would expect in a handbook. This section is well edited (by W. J. Karplus), with a consistent point of view, a single set of symbols and terms, and a large proportion of figures and tables to summarize facts at a glance.

The type font used in the body is, for a handbook, surprisingly large and readable. It is unfortunately marred by poorly chosen bold-face type: minor headings overshadow major headings, obscuring the structure of the text. Reference would be simplified if section headings were repeated at the top of each page, as is the usual practice in handbooks.

It is to be hoped that the editors and publishers will have the opportunity to correct shortcomings in a future edition. PerPROCEEDINGS OF THE IRE

simultaneous transmission of L and X band and S and X band emissions through the electron plasma generated in the Hot Shot tunnel. The receiving antennas were immersed in the plasma inside the tunnel. The observed results are compared with theoretically computed values and the limitations of this technique are outlined. The experimental results tend to verify the theoretical estimates of the deleterious effects of the hypervelocity sheath. These results, along with Fastax movie records of the glow from the shock, will be presented.

53.5. Ultra-Low Frequency Atmospherics

HERBERT KÖNIG, Electro-Physical Inst., College of Tech., Munich, Germany.

It was possible to establish by measurement that in the frequency range from 0.5 cps to 25 cps, electrical signals are present in the atmosphere whose origin is not trivial.

At that occasion the existence of electromagnetic oscillations was proved with a frequency of about 9 cps. The origin of these signals probably lies in flashes of lightning which excite the resonator earth-ionosphere to oscillate in its basic frequency, thus creating signals whose conditions for spreading are very favorable.

Besides, various electrical field variations of a more local character were observed. These signals probably were connected with certain weather conditions.

In addition to an occurrence during sunrise which probably was caused by processes during the development of the ionosphere in its daily structure, the measurements showed a few other electrical and magnetic special forms whose origin is unknown.

53.6. Ray Tracing for Whistler-Mode Signals at Low Frequencies

E. R. SCHMERLING, Pennsylvania
State University, State College, Pa.;
R. GOERSS AND S. MILUSCHEWA,
RCA, Hightstown, N. J.; AND P.
HERTZLER AND I. PIKUS,
RCA, Camden, N. J.

A number of whistler-mode ray-paths have been traced in an IBM 650 computer using Haselgrove's equations. A frequency of 5 kc was taken at a geomagnetic latitude of 50° for various initial propagation angles. A centered dipole was used for the earth's field, and a simple ionspheric electronic-density model based on Seddon's (1957) composite curve. The purpose of the work was to examine the path spreading as a function of initial angle, not to obtain the exact conjugate points, so that the simple model was considered adequate. This problem has a special bearing on second-hop signals and satellite-originated signals, since signals originating on the ground are not expected to have a large spread of angle at heights of the order 100 km. A decided shift of the downcoming rays towards the equator was found, and a large spread of the order 1000 km for the angular spectrum.

SESSION 54*

Thurs.

2:30-5:00 P.M.

Coliseum Morse Hall

WAVEFORM ANALYSIS AND RANDOM VARIABLES

Chairman: MICHAEL DI TORO, Polytechnic Res. and Dev. Co., Brooklyn, N. Y.

54.1. A Time-Compressor Using Magnetostrictive Delay Lines

S. J. MEYERS, L. ROSENBERG, AND A. ROTHBART, *ITT Labs.*, *Nutley*, N. J.

Waveform analysis of low-frequency signals may be simplified by storing a given time segment of the signal, and then transferring this segment to a higher frequency band. The stored signal, which may be read out repeatedly many times faster than the speed of original storage, is transformed into a timecompressed repetitive waveform. It is then available for analysis during a time interval equal to that of the input signal.

This paper will discuss the design principles and test results of an experimental time-compressor which used magnetostrictive delay lines as the storage mechanism. The signal segment, 1/20 second duration, was bandwidth limited to 1500 cps. Within the compressor, pulse code modulation (PCM) was used to transmit eight analog levels in a reflected binary code. The low-frequency sampling rate was 4 kc and the compression factor was 200.

54.2. Utilization of the Quadrature Functions as a Unique Approach to Electronic Filter Design

HENRY PARIS, Rixon Electronics, Inc., Silver Spring, Md.

This paper describes a new approach to electronic filter design used to obtain a tunable band-pass filter in the audio spectrum. The method consists of utilization of the quadrature functions described by Dr. D. K. Weaver in the December, 1956 Single Sideband Issue and employs modulation, frequency translation, low-pass filtering, and phase cancellation to achieve a band-pass filter, with a variable center frequency, and a translated output range. The quadrature function theory is described in detail, both diagramatically and mathematically, through the use of slides.

A few of the obvious applications of the system are discussed; however, the presentation is in as broad a manner as possible to eliminate restriction of the design approach to the present Rixon application and to enhance stimulation of ideas for improvements and new applications for the quadrature function design approach.

54.3. A Magnetostrictive-Filter Random Wave Analyzer

RICHARD BOYNTON, MB Electronics, New Haven, Conn.

* Sponsored by the Professional Group on Instrumentation. To be published in Part 9 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

A new type of Audio Frequency Random Wave Analyzer presented in this paper contains two desirable features previously only available independently: rapid analysis time and narrow bandwidth scanning. Utilizing a parallel bank of narrow-band magnetostrictive rod filters and an unique solid-state circuit to translate each aperiodic signal filter output to a de level, this instrument permits measurement of stationary sources as they occur. Currently available single-filter sweep frequency analyzers offer scanning bandwidths as narrow as 2 cps but require over half an hour to analyze a single random input. Increasing this sweep rate necessitates wider bandwidth scanning, resulting in an erroneous smoothing out of peaks and notches in the spectral density curve.

54.4. A Numerical Method for Determining the Vibration Acceleration Density Directly from the Sinusoidal XY Plot

W. REICH AND M. SCHNEE, American Bosch Arma Corp.,

Hempstead, N. Y.

The acceleration density at a point on the vibration table, fixture or specimen may be obtained by processing a tape recording of the random acceleration at that point with an automatic wave analyzer. This paper describes a method to obtain the same information using only the standard equipment on the MB T88 Random Motion Console.

The advantages of this method are:

- 1) The tape recorder and wave analyzer may be used elsewhere.
- Time required is only twenty minutes compared to two hours using the tape and analyzer.
- The acceleration density may be predicted without actually subjecting the equipment to random vibration.

54.5. A New Approach to Random Vibration Control Instrumentation

WILLIAM W. CALDWELL, Ling-Altec Electronics, Inc., Anaheim, Calif.

Equalization, applied to random vibration testing, is the process of generating a continuous function, the complement of that of the shaker plus load measured at the shaker mounting point, so that the net frequency response of the vibration test system is flat. The new approach to the problem limits with a new equalizer the energy in each of a number of narrow frequency bands covering the entire vibration test spectrum. The output signal of each frequency band can be controlled over a wide range, then recombined creating a shaped, equalized spectrum.

A primary advantage of the new equalization system is to allow continuous display of the average acceleration power spectral density over each frequency band during the random vibration test with a monitor containing a filter set identical to those of the equalizer system. Changes in the spectrum during test, which occur because of changes in ambient temperatures, mounting bolt tension, and the like, can be quickly identified and compensated at the equalizer. the actual measured path loss to these points. Four representative frequencies were chosen in the range 1200 to 9000 mc for the actual path measurements. A standard mobile Army surveillance type radar equipment operating at 1300 mc was used in making the map overlays. An average of twenty minutes is required to produce a map overlay. A correlation has to be established between the system gain of the vehicular equipment and system gain of the radar. This means that in general the system gain of the radar must be reduced in order to match the system gain of the communication equipment. This is most simply done by adjusting a calibrated gain control on the radar receiver. The adjustment must take into consideration the type terrain in the area surveyed. Illustration of map overlays will be shown in various terrain situations for distances up to ten miles.

52.3. Cryptographic Signaling Applied to Radio Communication Circuits

OWEN E. THOMPSON, Secode Corp., San Francisco, Calif.

The primary concern of this paper is a discussion of a system wherein multiple, *secrel* messages may be delivered to any one of many receiving locations on a radio circuit.

Present day one-way signaling (paging) systems will be discussed sufficiently to develop the historical background and establish the need for a system with both secrecy and multiple message capacity. The degree of secrecy of this system is inherently very high and will be discussed in detail. The basic system code capacity of 31 discrete messages, at any one of many receivers (actual capacity of system is many thousands of receivers) will be discussed.

Equipment description will be facilitated by display of actual working units. Part of the system employs standard Secode multiple function decoders of the same type that have been in industry-wide use for the past few years. Other portions of the system that are of recent development will receive the major portion of time allotted to equipment description,

In addition to system concept and technique of implementation, specific types of application will be discussed to illustrate the wide field of need that will be satisfied by this system.

52.4. Highway Alert Radio

E. A. HANYSZ, General Motors Corp., Warren, Mich.

The author's laboratory recently released the news of the development of a new roadsideto-vehicle communications system utilizing netic induction radiation. The purpose of this system is to introduce a voice message into the passing vehicle regarding road conditions, detours, exists and emergency situations directly ahead so that the driver is alerted to react wisely. Reports received from highway and traffic engineers seeing this system have been glowing and it appears that such communications will find widespread acceptance in the near future.

This paper discusses the development of portable transistorized transmitters and receivers, the problems attendant in modulating a sub-10-kc carrier at audio frequencies, temperature stabilization, antennas and patterns, and circuits for muting the standard car radio. The author is also entertaining the idea of bringing a complete system with him (since it is portable) and demonstrating it to interested parties.

52.5. A New Colinear Antenna Array

A. H. SECORD AND W. V. TILSTON, Sinclair Radio Labs., Ltd., Toronto, Ont., Canada

In order to serve ever-expanding communities it has become necessary in recent years to develop omnidirectional antennas with higher and higher gains. These antennas serve the need for the mobile base station and for repeater stations,

This paper describes the application of a special phase changer which when periodically distributed along a cylindrical conductor controls the phase of the radiating currents and so concentrates the radiation in a plane, perpendicular to the axis of the cylindrical conductor.

SESSION 53*

Thurs.

2:30-5:00 P.M.

Coliseum Marconi Hall

ANTENNA AND PROPAGATION PROBLEMS

Chairman: JACK HERBSTREIT, Natl. Bureau of Standards, Boulder, Colo.

53.1. Spiral Antenna Systems

R. BAWER AND J. J. WOLFE, Aero Geo Astro Corp., Alexandria, Va.

The basic purpose of this paper is four-fold:

- To place the spiral antenna in its proper perspective with regard to what may be reasonably expected from these antennas in practical situations.
- To present experimental data compiled from many antennas operating under a wide variety of conditions.
- To interpret these results in light of the band theory.
- To emphasize those aspects of parameters which significantly affect the operation of the antenna.

All of the spirals described have capability for accurately scaling from one frequency band to another. None of the spirals are plagued by the beam lobing with frequency and polarization; all spirals have zero boresight error. Excellent circularity is obtained not only on the peak of the beam but also throughout the spiral beam to the tenth power point. The data are presented to indicate how these units are superior to existing designs and how they can be easily reproduced by the antenna designer interested in an element covering 2 to 1 frequency band with superior pattern, polarization and impedance performance.

53.2. A Monopulse Cassegrainian Antenna

L. SCHWARTZMAN AND R. W. MAR-TIN, Sperry Gyroscope Co., Great Neck, N. Y.

The Cassegrainian antenna has become increasingly useful for radar systems application. A summary of the design considerations employed in selecting a Cassegrainian antenna. rather than a conventional antenna with the feed at the focal point, are presented. The application of the Cassegrainian antenna described here is for monopulse systems. For a chosen illumination of the secondary aperture, greater latitude in the choice of the actual feed horn dimensions is afforded by a Cassegrainian antenna. Design equations describing the basic parameters are included. The range- and angular-sensitivity functions which are of prime importance in a monopulse design are developed. Curves depicting their behavior as a function of horn size and F/D ratio are included. The near-field problem which arises if the main reflector is substantially less than 100 λ , is discussed along with methods of eliminating these effects. Experimental data are presented and compared to the theoretical predictions.

53.3. Power-Handling Capability of Antennas at High Altitude

W. E. SCHARFMAN AND T. MORITA, Stanford Res. Inst., Menlo Park, Calif.

The factors influencing the power-handling capability of antennas at high altitude are considered in this paper. The physical mechanism involved, including the roles of attachment, free diffusion, ambipolar diffusion, and nonuniform field distribution in the breakdown process, is qualitatively described. These factors are illustrated by breakdown curves for various antenna configurations under both CW and pulse conditions. Normalized data, which are useful for estimating breakdown fields when the conditions for scaling are fulfilled, are presented. Methods are then considered for increasing the power-handling capability, and typical results are given showing the increase in power that can be achieved.

The effect of missile environment on breakdown characteristics is discussed, and an experiment that involves artificially introducing ionization near the surface of the antenna is described.

53.4. Propagation Measurements in Shock-Ionized Media

D. E. SUKHIA AND G. H. HAMPTON, The Martin Co., Baltimore, Md.

This paper presents the results of propagation experiments conducted in Hot Shot II, **a** hypersonic wind tunnel at AEDC in Tullahoma, Tenn. The experiments consisted of

^{*} Sponsored by the Professional Group on Antennas and Propagation. To be published in Part 1 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

51.2. The Role of Multipurpose Automatic Test Systems in Testing Integrated ABNMGS Systems

DR. IBRAHIM H. RUBAH, IBM Federal Systems Div., Owego, N. Y.

To assure the continuing satisfactory opera-

tion of a complex bombing-navigation and missile guidance system, extensive efforts must be expended in the testing process, which depends heavily upon the test system. In this paper, the testing process and test systems are treated in analytical fashion. The test system is divided into basic functional elements that are universal to all test systems. These functional elements, or subsystems, are then integrated into a *theoretical automatic multipurpose test system*. Boundary conditions imposed on this general system would lead to any special purpose test system.

51.3. Selecting the Optimum Test Interval for Static Alert Systems

F. PAULSEN AND L. MAST, Packard Bell Electronics, Los Angeles, Calif.

A method was found for determining the best frequency of tests applied to static alert systems such as missiles. Two mathematical models are established in which the best test frequency is expressed in terms of cost, failure rates, duration of test, and down time due to maintenance or repair. Conclusions include that:

- The most economical arrangement requires that the test frequency be less than the test frequency which gives maximum operational readiness.
- In general, different test frequencies are appropriate for different tests. Therefore, the current practice of establishing a "test sequence" and performing all tests at the same test intervals should be critically examined.

51.4. Rapid Detection of Coherent Signals in Noise

R. J. METZ, J. M. WALKER, AND N. L. WEINBERG, Westinghouse Electric Corp., Air Arm Div., Baltimore, Md.

A method is described of detecting coherent signals in noise based on noncoherent integration allowing rapid signal detection in a multichannel receiving system. This is accomplished by time-sharing the outputs of each of the channels with a common detection criterion. To eliminate problems incident with time sharing a large number of conventional RC integrators a mechanization is described which uses square loop magnetic cores to perform all required functions. The noncoherent integration function is performed by the accumulating action of the core.

The results of this investigation are presented as curves of probability of detection vs signal-to-noise ratio. An evaluation is made between magnetic core detection and RC detection.

51.5. Determination of Repetition Frequencies of Intermixed Pulse Trains

RONALD J. KERN, RCA, Camden, N. J.

The problem of determining the pulse repetition frequencies contained in a composite signal received from a large group of pulsed transmitters is discussed. This problem, which arises in connection with military reconnaissance receivers used to determine the PRF of enemy radars, is complicated by the wide range of PRF's encountered, the high degree of measurement precision required, and the unknown number of transmitters present.

Two methods of PRF determination under the stated conditions are presented in this paper. The first method involves time domain autocorrelation while the second method consists of a network of cascaded delay lines, the delay of each being a function of position in the network. Mathematical analysis and practical mechanization techniques are presented.

51.6. Coherent Enhancer for Pulse Radar Applications

E. BROOKNER AND J. FLINK, Federal Scientific Corp., New York, N. Y.

The ideal matched filter with many delay lines is difficult to implement. Thus, rather than use the ideal filter, a technique which approximates the ideal matched filter was devised consisting of a single delay line loop which is much simpler to implement. In addition to extending the range, the matched filter will provide Doppler velocity information and also has the features of an MTI.

The matched filter is compared against the more common video integrator. It is shown that the range extension capabilities of the matched filter are superior.

SESSION 52*

Thurs.

2:30-5:00 P.M.

Coliseum Faraday Hall

VEHICULAR COMMUNICATIONS

Chairman: R. P. GIFFORD, General Electric Co., Lynchburg, Va.

52.1. Past and Future Techniques of Vehicular Communications

E. W. CHAPIN, FCC, Laurel, Md.

A résumé will be made of the development of vehicular communications, starting with the pre-electric age and pre-wireless age. The general history of the early wireless communication to moving vehicles will follow and comparisons will be made between the characteristics of the early radio-telephone equipments and later types, with respect to both the improvements which increased the utility of the equipments and those improvements which permitted more efficient use of the radio frequency spectrum. Tabulations will show the changed technical characteristics over the years. These characteristics will cover: crystal harmonics, spurious harmonic problems, frequency stability, and bandwidth problems.

Possible approaches to obtain still more efficient utilization will be discussed as well as considering newly available components and equipments which may increase the need for more facilities. For example, the availability of really portable transistorized equipment may tend to develop a whole new field of communication similar in scope to the land-mobile comnunications systems presently so widely established. In some instances, this new expansion will undoubtedly be integrated with present systems, but, in other cases, combination with landmobile operation may not be involved.

52.2. Radio Coverage—Area Survey —Instrumentation Research

C. E. SHARP AND R. E. LACY, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.

The rapid establishment of effective shortrange radio communications is of the utmost importance in modern Army operations. Due to the over-occupancy of the lower frequencies, the need to use the UHF and SHF bands for short-range communications has become imperative. The higher frequency bands also have the advantage that vehicular antennas of small physical size can have high gain and directivity.

Transmission between terminals in the higher frequency bands is normally limited to radio line-of-sight. The usual methods for determining whether transmission is possible between selected terminals are time consuming and inconclusive. Path loss prediction methods have been devised for various types of irregular terrain; while these methods predict the average expected value, however, for any given path the deviation may be as much as plus or minus 30 db.

Investigations at the U.S. Army Signal Research and Development Laboratory of survey system using a combination of radar and photographic techniques have indicated strong possibilities for a rapid and effective means of determining all possible satisfactory shortrange communication paths from any given location with an antenna height above ground equal to the average vehicular antenna. The survey system involves making a composite photograph of a mobile radar plan position indicator and map of the area centered at a given location with careful attention to orientation and scale factors. Areas on the composite photograph from which terrain backscatter is evident, with some simple considerations, can readily be determined as satisfactory terminals for communication circuits.

Interest in this survey system stemmed from the need for rapid siting of point-to-point vehicular radio communications employing directive antennas. This interest has led to an extensive experimental investigation of numerous backscatter map overlays in all conceivable types of terrain. The purpose of this investigation was to establish definitely for all types of terrain the correlation which exists between the terrain backscatter return points and

^{*} Sponsored by the Professional Group on Vehicular Communications. To be published in Part 7 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

50.2. A Pulsed Plasma Mechanism for Propulsion in Space

P. M. MOSTOV, J. L. NEURINGER, AND D. S. RIGNEY, *Republic* Aviation Corp., Farmingdale, N. Y.

A variable wall-resistance as well as a constant source-resistance are included in the analysis of a plasma accelerator. Coupled nonlinear equations involving 7 parameters are derived. This formulation is more realistic than Artsimovitch's treatment (source, wallresistance zero) and Shock's approximate periodic mode treatment (wall-resistance zero). Curves of position, velocity, efficiency, utilization, instantaneous efficiency, voltage, current, and instantaneous frequency are given. Periodic and aperiodic modes, as well as modeswitching and switch-backs are shown possible. In one typical case, including source-resistance lowers efficiency from 100 to 30 per cent; including wall-resistance further lowers it to 22 per cent. Practical optimization of utilization by terminating at local peaks can frequently be achieved with reasonable tube lengths.

50.3. Design Considerations of Television Satellite Reconnaissance Systems

R. L. ZASTROW AND D. J. RITCHIE, Bendix Aviation Corp., Detroit, Mich.

The following applications of satellite television reconnaissance are briefly considered; weather reconnaissance, shipping and ice patrol forestry surveys, mapping and military reconnaissance. The coverage, frequency of observation, grey scale rendition, and resolution requirements for each application are discussed.

Television systems using vidicons and image orthicons are compared. The sensitivity, resolution, dynamic range and spectral response of the vidicon and image orthicon are discussed. The design parameters of a television system (frame rate, picture format, band-pass, number of lines per raster) are analyzed and related to system requirements for resolution and coverage. The influence of optical design parameters such as field-of-view, distortion, focal length and relative aperture is analyzed in terms of system requirements.

The information content of a television picture is defined. The power and bandwidth required to transmit the video signal to a ground station is analyzed for different methods of encoding the telemetry carrier. The use of a tape recorder aboard the satellite, and the problems of obtaining sufficient bandwidth and recording capacity are discussed.

50.4. Scanning Methods for Satellite-Borne Radars

A. ROSENFELD AND O. LOWENSCHUSS, Budd Lewyt Electronics, Inc., Long Island City, N. Y.

This paper is concerned with scanning methods suitable for use in earthward-looking satellite-borne search radar systems. Particular attention is given to the case of a pencil-beam radar borne by a polar-orbit satellite. The corresponding airborne case is analyzed, and the need for special methods in the space-borne case is explained. Methods are then considered which make maximum use of the translational and rotational motions of the satellite to aid in effecting the scanning process. The advantages and disadvantages of such methods are examined in some detail.

50.5. A Study of Natural Electromagnetic Phenomena for Space Navigation

R. G. FRANKLIN AND D. L. BIRX, The Franklin Inst., Philadelphia, Pa.

A study has been carried out for the United States Air Force investigating the use of natural electromagnetic radiation in the space environment for navigational purposes. Radiations from the sun, the stars, and interstellar space in both the visible and radio frequency portions of the spectrum, and cosmic rays have been investigated.

Emphasis has been placed on the measurement of velocity in space utilizing the Doppler phenomenon. Equipment and techniques useful in deriving velocity information from Doppler shift measurements are described and figures for expected accuracy are derived. Other passive techniques having possible application to space navigation such as the measurement of total solar radiation and solar diameter are touched on briefly.

SESSION 51*

Thurs.

2:30-5:00 P.M.

Waldorf-Astoria Sert Room

CHECK-OUT INSTRUMENTA-TION AND CIRCUITRY

Chairman: DR. J. Q. BRANTLEY, Radiation, Inc., Orlando, Fla.

51.1. Trends in Complex Weapon System Checkout

F. C. COREY, Nortronics Div., Northrop Corp., Hawthorne, Calif.

Methods of checkout that will be used in testing the electronic portions of complex weapons systems and space vehicles from 1963 to 1970 are described and key factors affecting the choice of manually operated, semiautomatic, or completely automatic equipments are discussed. Relative advantages of externally programmed test equipment controlled by tape or cards vs internally programmed test equipment controlled by digital computers, magnetic memories or stepper switches are evaluated. Methods of mechanizing equipment to give more than "go, no-go" answers are described and a high-speed, computer-programmed digital system using magnetic drum memory storage is proposed.

* Sponsored by the Professional Group on Military Electronics. To be published in Part 8 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

49.4. Application of Synthesis Techniques to Electronic Circuit Design

FRANKLIN II. BLECHER, Bell Telephone Labs., Inc., Murray Hill, N. J.

During the past decade, considerable progress has been made in the field of RC-active network synthesis. Techniques are now available for realizing a wide range of driving point and transfer characteristics. This work is extremely important since it defines the region of theoretical possibilities. The question arises, though, as to how useful this synthesis theory is for practical circuit design. Active circuit elements such as negative impedance converters and feedback amplifiers using solidstate devices can be designed to have the same stability as the best passive circuit elements. Unfortunately, however, the nature of present synthesis techniques is such that the resulting RC-active networks are more sensitive to both passive and active circuit element variations than in the case of passive RLC networks.

Some of the more useful active network synthesis techniques will be reviewed and compared from the points of view of generality, sensitivity, and numbers of passive and active circuit elements required to realize a given network function. It will be shown that at frequencies below about 30 kc, RC active networks are competitive with all other methods of filter design. Finally, two of the synthesis techniques are used to realize a twelfth-order equal-ripple Tchebycheff parameter band-pass filter centered at 18 kc.

SESSION 50*

Thurs.

Waldorf-Astoria

2:30-5:00 P.M.

Jade Room

SPACE ELECTRONICS

Chairman: PAUL POLISHUK, Wright Air Dev. Center, Wright-Patterson AFB, Ohio

50.1. A Broad-Band Spherical Satellite Antenna

H. B. RIBLET, Johns Hopkins University, Silver Spring, Md.

The design criteria and results of the design of a broad-band spherical antenna are discussed. The parameters of design and their effect on radiation patterns are briefly investigated. The particular antenna was developed from an equiangular spiral slot plotted on a plane and then projected on the surface of a sphere. Attention is given to problems of isolation and matching when the antenna is used for multifrequency operation. The particular antenna discussed is being used in the Transit satellite program.

^{*} Sponsored by the Professional Group on Space Electronics and Telemetry. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SESSION 48*

Thurs.

2:30-5:00 P.M.

Waldorf-Astoria Starlight Roof

ELECTRONIC COMPUTERS

Chairman: JOSEPH J. EACHUS, Minneapolis-Honeywell Regulator Co., Newton, Mass.

48.1. An On-Line Solid-State Analog Computer for Automatic Gas Flow Compensation

F. P. SIMMONS, Link Aviation, Inc., Binghamton, N. Y.

A completely automatic, solid-state, analog computer is described for continuously solving the basic equation used to calculate quantity rate of gas flow through an orifice for accounting and dispatching purposes. The computer is unique in that it performs multiplication, division, and square root extraction using only two transistorized multipliers in the system. An analysis of the parameters involved in the basic equation is included, which shows that specific gravity, in most natural gas installations, may be considered as a constant and contribute less error in the calculated rate of flow than would be contributed by considering it an independent variable. The paper includes an error analysis of the system along with the scaling method used to program the equation through the computer.

48.2. Very High Density Digital Magnetic Recording

DONALD E. KILLEN, Oliver-Shepherd Industries, Inc., Nutley, N. J.

Due to greatly improved heads and media, and the amplitude discrimination and restricted bandwidth of phase modulation techniques, 3000 bits per inch on tape is now possible. Operation is reliable near the noise level. Noncontact densities approach 500 bits per inch, due also to special low-loss drum and disk coating techniques. Other recording methods, (e.g., RZ, NRZ) rely on high signalto-noise ratio, and are therefore not compatible with high information density. A systems approach is tantamount to achieving high density; heads, media, and electronics must be carefully integrated. One proposed tape system would achieve a 1-mc character transfer rate at moderate velocity, for direct input to computer without buffering.

48.3. A Tunnel Diode Tenth Microsecond Memory

M. M. KAUFMAN, RCA, Camden, N. J.

A resistor load line placed across a tunnel diode (negative resistance diode) produces a two stable state storage unit. There have been tunnel diodes made to oscillate above 3 kmc indicating the high-speed capabilities of the above storage unit.

A type of tunnel diode memory cell believed to be practical for a large (approximately 10^b bits) memory has been developed. The memory cell is a baseband destructive "read" type with a resistor loaded tunnel diode and a transformer coupled output. The cell has been chosen on the basis of its simplicity and relatively large magnitude of output signal. The large magnitude of output is achieved with the transformer coupling and is very significant at the tenth microsecond memory speeds.

Packaging techniques have been developed for the memory cells which produce large packing densities and maintain the proper electrical properties at a reasonable cost.

A theoretical and laboratory study has been made of memory cell and distributed memory matrix parameters with an attempt to specify optimum parameters.

An 8×8 bit plane with the over-all dimensions and selection lines dummy loaded to simulate a 32×32 bit plane has been built and successfully tested in the laboratory.

48.4. Automatic System and Logical Design Techniques Used on the RW-33 Computer System

T. A. CONNOLLY, Thompson-Ramo-Wooldridge Inc., Los Angeles, Calif.

Techniques to improve and aid logical and system design are discussed. These techniques shorten the deve opment cycle by verifying the design before the equipment is assembled. The method eliminates many costly logical revisions and decreases system checkout and integration time. The main technique discussed is the logical simulation package. Secondly, integration of the logical and instruction simulators which improves both logical and system integration is discussed.

48.5. Logical Design Features of the Larc System

W. F. SCHMITT AND L. F. HARRISON, Remington Rand Univac, Philadelphia, Pa.

The logical structures and operations of the several Univac[®]-Larc system elements are described. The overlapped control cycle and major arithmetical operations of the computing unit are outlined. The memory unit logic and system synchronization are described. The Univac-Larc Input-Output Processor is described briefly with particular emphasis upon the input-output circuitry and the input-output dispatcher. A brief account is given of the logical elements employed in the design of the system.

SESSION 49*

2:30-5:00 P.M.

Thurs.

Waldorf-Astoria Astor Gallery

* Sponsored by the Professional Group on Circuit Theory. To be published in Part 2 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SYMPOSIUM ON A DECADE OF PROGRESS IN NETWORK THEORY

Chairman: H. J. CARLIN, Polytechnic Inst. of Brooklyn, Brooklyn, N. Y.

49.1. Graph Theory and Electric Networks II

FRANK HARARY, University of Michigan, Ann Arbor, Mich.

Some of the open problems mentioned in the article, "Graph Theory and Electric Networks" (IRE TRANSACTIONS ON CIRCUIT THEORY, vol. CT-6, pp. 95 109, March, 1959) have since been solved by various authors. Other results which exploit graph theoretic methods for applications to the combinatorial aspects of electrical network theory have recently appeared. These articles deal with the following topics: 1) Boolean functions and synthesis problems, 2) the spanning trees of a network, 3) the realizability of cut set matrices, 4) the consistency of precedence matrices, and 5) finite automata. A progress report will be presented emphasizing latest developments.

49.2. Physical Realizability Criteria

D. YOULA, Polytechnic Inst. of Brooklyn, Brooklyn, N. Y.

A rigorous theory of linear, passive, timeinvariant *n*-ports is surveyed from an axiomatic point of view. The notions of linearity, passivity, time-invariance, and causality are defined precisely for the most general operator which maps *n*-dimensional vector functions into *n*-dimensional vector functions. One main result is that with the exception of certain pathological cases, linearity and passivity imply causality. Moreover the causality postulate plays absolutely no role in a theory of linear, passive networks.

Some of the physical implications of Maxwell's equations when applied to dielectric media are also discussed, particular emphasis being placed on the construction of nonreciprocal structures.

Lastly an algebraic treatment of lumped active metworks is outlined which includes both the analysis and synthesis problems.

49.3. Some Properties of Time Varying Networks

J. M. MANLEY, Bell Telephone Labs., Inc., Murray Hill, N. J.

In this paper, the general theory of reactance amplifiers is exemplified, with attention centered on three aspects. 1) Why may gain be obtained from a nonlinear reactance and not from a nonlinear resistance? 2) Where there is gain and feedback, the possibility of oscillation exists. But the situation is more complex than it is with vacuum tube amplifiers because the feedback path is through the modulator so that an external impedance at one frequency may be reflected as an impedance at another frequency. 3) The particular case of three frequencies, e.g., p, q, p-q, is calculated for large signals. The phenomena of regeneration and oscillation, with subharmonic oscillation as a special case, are studied and illuminated.

^{*} Sponsored by the Professional Group on Flectronic Computers, To be published in Part 2 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

46.3. Parasitic Spiral Arrays

1960

R. M. BROWN, JR. AND R. C. DOD-SON, U. S. Naval Res. Lab., Washington, D. C.

Arrays of circularly polarized elements have the property that their radiation patterns can be scanned by element rotation since rotation changes the element phase. This principle has been used in the design of two different arrays of Archimedean spirals, one a "parasitic lens array" and the other a "parasitic reflector array."

The lens consists of two arrays of spirals mounted on opposite sides of a conducting plane, each spiral on one side being connected by a transmission line to a corresponding spiral on the opposite side. A feed horn is used to illuminate one of the arrays; the energy received by this array passes through the ground plane via the transmission lines and is reradiated by the spirals on the far side. If the elements are rotated to particular orientations, the incident spherical wave can be focused, and further, the beam can be scanned if the spirals are rotated properly. The passive spiral reflector is a single array (using no transmission lines) mounted above a ground plane and, like the lens, is illuminated by a feed horn. It has the same focusing and scanning properties as the lens. Neither array is limited to circular polarization. With a combination of right- and left-hand elements, linear polarization can be used. Twenty-element lenses (2 by 10) and one hundred-element reflectors (10 by 10) have been built and the results have verified the predicted performance.

46.4. An Electromechanically Scannable Trough Waveguide Array

W. ROTMAN, AF Cambridge Res. Center, Bedford, Mass., AND A. MAESTRI, Melpar, Inc., Falls Church, Va.

This paper describes an electromechanical trough waveguide array which was designed and constructed to demonstrate the feasibility of a rapid scanning antenna utilizing a traveling-wave trough waveguide feed. The trough waveguide array was used to illuminate a section of a parabolic reflector. The antenna, a linear array, depends on several properties of the trough waveguide. Two methods are described which are capable of electromechanically swinging the fixed beam of a linear array based on the trough waveguide by changing the phase velocity in the guide. The two methods are:

- Rotation of symmetrical structures within the trough waveguide which have the quality of producing different values of guide wavelength in different orientations.
- Mechanically varying the height of periodic structures located on the top of the center fin of the trough waveguide.

Both methods were incorporated in short arrays with the latter also used in an array which is 10 feet long. Successful operations with scan angles in excess of 20° have been obtained from both methods.

Experimental data are presented on the 10foot array feeding a parabolic cylinder. SESSION 47*

Thurs. 10:00 A.M.-12:30 P.M.

Coliseum Morse Hall

MAGNETIC RECORDING

Chairman: MARVIN CAMRAS, Armour Res. Foundation, Chicago, Ill.

47.1. The Effects of Track Width in Magnetic Recording

D. F. ELDRIDGE AND A. BAABA, Ampex Corp., Redwood

City, Calif.

The effects of track width on various performance characteristics have been measured over a wide range of widths. Signal level, noise, and signal-to-noise ratio were determined for track widths from 1 mil to 92 mils. The effects of crosstalk, tape guiding, and actual recorded track width vs head width are described. The experimental data are in good agreement with theory and no serious practical limitations on the use of very narrow tracks are discovered. High density audio and pulse recordings were made without difficulty. Digital bit densities of one million per square inch, and above, are shown to be possible.

47.2. Erased Carrier Recording

WILLIAM J. MURPHV, Oliver-Shepherd Industries, Inc., Nutley, N. J.

There are many applications in which it is desirable to record data with a spectrum extending down to dc. One relatively unknown method, called erased carrier or amplitude modulation recording, will be the subject of this paper. A theory of operation will be developed as well as an explanation of the process. Approximate results for a typical system will also be presented. Advantages and disadvantages of such a system will be discussed, as well as some suggested applications.

47.3. Reliability and Drop-Out Studies for Long-Playing Loops

AL WILSON, Precision Instrument Co., San Carlos, Calif.

Although the tape loop has been used extensively for repeated playback analysis of selected portions of instrumentation data, there is little complete information available which reflects its performance characteristics after long-term repetitive playing. Because of increased interest in the tape loop technique and the belief that it would become a more useful laboratory tool once operational limitations were known, an exhaustive study was conducted. The effects of basic design, length of loop, tape speed, and type of tape were considered. The reliability under various conditions was measured by counting the number of "drop-outs" (instantaneous loss of signal) on the tape after a specified period of operation. The results of the study are, for the most part, presented in the form of graphs, charts, and statistics.

47.4. Digital Magnetic Recording with High Density Using Double Transition Method

ANDREW GABOR, Potter Instrument Co., Inc.,

Plainview, N. Y.

The solution of three major problems in high-density multichannel recording, such as pulse crowding, interchannel time displacement and information drop-out, is discussed.

After a study of these limiting factors, a write-read system with many favorable features is described in detail and results of extensive experimental investigation are given. It is shown that the use of the described techniques offers a storage capacity of 24,000 bits per square inch of tape (1500 bits per inch longitudinal and 16 channels per inch lateral packing density) combined with a character rate of 225 kc at 150 ips tape speed.

Reliability is greatly in advance of conventional parallel recording with compatable density. Drop-out rates better than one bit in 500 million have been observed.

47.5. Automatic Error Detection Equipment for Digital Tape Recorders

G. J. SLUSARCHYK, T. D. RADWAY, AND P. HELLER, Airborne Instruments Lab., Mineola, N. Y.

This paper describes a transistorized electronic device for examining the performance of a high packing density, start-stop, spaceborne digital tape recorder.

The equipment described automatically writes preset words at varying rates to full recorder capacity. Upon playback, the recorder's output is electronically examined for skew, effects of wow and flutter, word spacing, word lengths, and word bit content. The error detection equipment provides word error figures for the tape recorder in a readily accessible form without the requirement for additional detailed data evaluation.

This equipment utilizes basic diode logic techniques and digital modules such as transistorized flip-flops and monoshots. Reliability of the error detector circuitry is achieved through self-checking features.

^{*} Sponsored by the Professional Group on Instrumentation. To be published in Part 9 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

44.5. A Digital Data Handling System for Real-Time Computation on the Atlantic Missile Range

M. P. FALLS, RCA Service Co., Patrick AFB, Fla., AND T. A. CHRISTIE, JR., Stanford University, Stanford, Calif.

Procurement of improved tracking radar and digital computing devices at the Atlantic Missile Range required development of new high-speed data handling and transmission equipment which would link these devices utilizing existing communication facilities.

The essence of the newly developed equipment is a data transmission system capable of *reliable* 3000-band operation over a toll-quality carrier volce-channel. Quaternary, pulses are utilized to conserve bandwidth, and reliable FM is the transmitting medium.

The equipment transfers data from a highprecision CW tracking radar, known as AZUSA to an 1BM 709 computer used in making realtime calculations of impact and apogee predictions of ballistic missiles. Operating performance data for this type of equipment over typical communications facilities are included.

SESSION 45*

10:00 A.M.-12:30 P.M.

Thurs.

Coliseum Faraday Hall

HUMAN FACTORS IN ELECTRONICS

Chairman: ROBERT R. RIESZ, Bell Telephone Labs., Inc., Murray Hill, N. J.

45.1. Coding Equipment for Ease of Maintenance

J. H. ELV, Dunlap Associates, Stamford, Conn.

The purpose of this study was to develop recommendations concerning information to be displayed on prime electronic equipment. Specifically, it was recommended that the following information be displayed (visual aids will be shown to demonstrate each recommendation):

- 1) Designation of functional groups of equipment,
- 2) Identification of signal paths,
- Identification of test points and indication, when appropriate, of sequence in which they should be used, and
- Presentation of historical data displaying periodic readings taken at each test point when equipment was operating satisfactorily.

45.2. The Replaceable Component: Key to Maintainability

ROBERT B. MILLER, IBM Corp., Poughkcepsie, N. Y.

Maintenance consists of checking, adjusting, troubleshooting, replacing, and sometimes repairing.

Replaceable packages or assemblies are the key to first-level maintainability because they determine: essential test points; contents of troubleshooting diagrams and service manuals; troubleshooting strategies and mean fault-location time; accessibility requirements; and nature and extent of training for first-level maintenance personnel.

This is especially true where the basic package is designed with a consistent level of maintenance knowledge requirements in mind. Updating of engineering change information can be simplified. In designing basic packages, therefore, maintainability considerations are worthwhile trade-offs against economy in manufacture and inventory.

45.3. A Procedure for Predicting Reliability of Man-Machine Systems

P. C. BERRY AND J. J. WULFF, Psychological Research Associates, Inc., Arlington, Va.

In complex man-machine systems output is not dependent upon the performance of every component since some serve to monitor others and to initiate back-up performance for failure. Such systems may be described as a network of contingent functionings, each function being classified as multiplicative, additive, or shunting. The description must encompass all functions, man and machine, operational and maintenance. Given the probability of adequate function of each, we may calculate the probability that enough will function adequately to produce system output. For human performance, we may calculate the probability of excessive variation from an average performance level by a method analogous to inspection by variables. Once a value of system reliability is obtained, the reliability of a subsystem may he evaluated by its proportionate effect upon total system reliability.

The procedures described were developed under contract with the Burroughs Corporation, Paoli, Pa., and are presented by permission.

45.4. A Method for Anticipating Human Factors Requirements in Manned Weapon Systems

MILTON A. GRODSKY, The Martin Co., Baltimore, Md.

The utility and need for human factors information early enough in a weapon system's development are discussed in terms of recommendations for equipment design and training and selection of operations personnel. Various methods of anticipating these requirements are given, particularly those concerned with missile and space systems. A critique of each of these methods is presented and the criteria for an optimum method are given.

A method employing the above criteria is presented and an example of its applicability to a space system is developed. The utility and accuracy of this method is also exploited.

SESSION 46*

Thurs. 10:00 A.M.-12:30 P.M.

Coliseum Marconi Hall

SCANNING ANTENNA ARRAYS

Chairman: CARLYLE J. SLETTEN, A F Cambridge Res. Center, Bedford, Mass.

Panel Members: J. RUZE, Radiation Engrg. Lab., Maynard, Mass; H. SHNITKIN, The W. L. Maxson Corp., New York, N. Y.; AND A. E. MARS-TON, Naval Res. Lab., Washington, D. C.

46.1. An Electronically Scanned Circular Antenna Array

H. P. NEFF AND J. D. TILLMAN, Elec. Engrg. Dept., University of Tennessee, Knoxville, Tenn.

A circular antenna array is described in this paper which has the following characteristics:

- Any azimuthal pattern can be obtained which can be represented by a truncated Fourier series.
- The main beam can be pointed to any azimuth angle.
- 3) The phase of the current in each element and the terminal impedance of each element does not depend on the direction of pointing.

These characteristics make it possible to connect amplifiers to each element, and to control the direction of pointing by varying the magnitude only of the amplifier output. The design of an array of this type and of the required amplifiers is described, and experimental confirmation is presented.

46.2. Multidirectional Antenna— A New Approach to Stacked Beams

JUDD BLASS, The W. L. Maxson Corp., New York, N. Y.

This paper describes a new approach to the design of stack beam antenna. The technique utilizes a multiplicity of linear arrays of directional couplers, the outputs of which are connected in series to the radiating elements of a linear array.

It is shown that the coupling between feedlines is at the sidelobe level if the phase distribution in the individual feedline results in a radiation beam divergence of more than one beamwidth. Thus the multidirectional characteristic is obtained from a single linear array such that, when coupled to N input connection, N independent beams are produced.

The design analysis and results of tests of a ten-foot X-band model are described.

^{*} Sponsored by Professional Group on Human Factors in Electronics. To be published in Part 10 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

^{*} Sponsored by the Professional Group on Antennas and Propagation. To be published in Part 1 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

43.3. Weather Radar Data Processing

O. LOWENSCHUSS, Budd Lewyt Electronics, Inc., Long Island City, N. Y.

This paper discusses the development of a radar data processor, used for encoding the intensity and height of the cloud cover over an extended area (44,000 square miles). The processor accepts video signals from a weather radar set, quantizes and averages the data (removing interference from external sources), and displays the data locally. In addition, the processor stores the data, and transmits them at high speed upon demand by the central data processor. The processor is intended for use in the Air Force Weather Observing and Forecasting System 433L, which contains a network of observation stations linked to a central data processing center by a high-speed wire transmission system.

43.4. A Building-Block Approach to Multipurpose Communication Equipment

L. G. FOBES AND J. E. MARTIN, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.; H. A. FRENCH, W. L. GLOMB, AND M. W. GREEN, ITT Labs., Nutley, N. J.

A design approach permitting functional flexibility with optimized performance in multipurpose, multichannel, communications equipment is discussed. Interchange of building block plug-in units would permit transmission to be tailored to 12, 24, 48, 72, 96, or 120 voice channels by PPM, PCM, or FDM in line-ofsight, over-the-horizon, or satellite communications systems in different frequency bands such as 2 or 4 kmc, in vehicular or fixed installations. Control of transmitted and received bandwidth by the use of Gaussian filters is discussed, as is the use of triode doublers to the upper S-band region, and IF combining for PCM troposcatter. Modular construction, conservative design, and applied human engineering, can all be combined to achieve a basic equipment design which can be readily modified to achieve the desired operational characteristics for a variety of military tactical situations.

43.5. An Integrated Approach to the Design of Mobile Tactical Electronic Systems

R. N. SKALWOLD AND M. N. SCHEIDERICH, *Rome, N. Y.*

This paper investigates the hypothesis that mobile military electronic systems can be drastically reduced in weight and volume for strategic and tactical mobility if a sufficiently broad systems viewpoint is taken.

The results of the study indicate that such an integrated approach can result in a complete radar system weighing only one third the weight of an extremely advanced lightweight radar system not completely integrated.

43.6. Electronic Equipment Weight and Volume Penalties to Flight Vehicles

WILLIAM V. WHITE, Collins Radio Co., Cedar Rapids, Iowa.

The existence of electronic equipment weight and volume penalties to high-performance flight vehicles is well known, but the means for evaluating the magnitudes of these penalties has not been completely defined. A means for penalty evaluation is presented for several types of aircraft. For existing aircraft, equipment weight and volume penalties are evaluated in relation to their effect upon range and endurance. For proposed flight vehicles, the penalties are evaluated in relation to their effect upon the over-all vehicle gross weight. Weight and volume penalties have been evaluated individually for both types of aircraft. Also, to permit more generalized interpretation, a combined weight and volume penalty analysis is presented. As a result of penalty evaluations, equipment planners and designers will be more aware of the effects of equipment weight and volume upon flight vehicles and will realize the ultimate benefits of equipment miniaturization efforts.

SESSION 44*

Thurs. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Empire Room

SATELLITE COMMUNICATIONS

Chairman: CHARLES II. DOERSAM, JR., Sperry Gyroscope Co., Great Neck, N. Y.

44.1. Radio Relaying by Reflection from the Sun

D. J. BLATTNER, RC.A Labs., Princeton, N. J.

The possibility of using radio reflection from the sun to provide a communication link supplementing that using moon reflections is investigated analytically. Reflection and absorption of an incident signal in the solar atmosphere, and thermal noise radiated by the sun, are considered together with transmitter capabilities to find the operating frequency giving optimum signal-to-noise ratio. The design principles for a receiver which can separate the signal from the noise are indicated. Finally, achievable signaling rates and reliability are calculated.

44.2. Active vs Passive Satellites for a Multistation Communication Network

L. POLLACK AND D. CAMPBELL, ITT Labs., Nutley, N. J.

The operational world-wide communication system will interlink many countries with widely different traffic requirements.

The use of space stations for relaying traffic in such a system poses the problem of selection of a passive or active satellite.

Among the problems discussed are the interference within the system and to other services, the power required vs message capacity, and the control of traffic.

44.3. Satellite Communication Problems and Solutions in Ground Station Design

W. L. GLOMB, ITT Labs., Nutley, N. J., AND W. TEETSEL, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.

The operational requirements of a low altitude "store and forward" satellite communication system are reviewed and some unique problems defined. Particular problems are reviewed and their solutions described.

In general, this typical satellite communication system is characterized by one station moving at high speed, by intermittent communications, by a limitation in weight, size, and complexity of the satellite with the corresponding burden in equipment and design ingenuity in the ground station, and by communication ranges considerably beyond those encountered on the earth's surface. Designs anticipating these characteristics have been evolved and have been applied to the communication system.

The derivation and application of these designs in the system are discussed in some detail. Their ultimate application in the Army's ACAN system is described.

44.4. Detail Design of an Operational Missile Voice Frequency Communications System

WARD S. CAYOT, Nortronics, Hawthorne, Calif.

This paper describes a detail design program to develop a voice frequency communication system for use in checking out Snark intercontinental missiles at a two-squadron missile strategic site.

The program is traced from predesign analyses of acoustics problems and questions of speech intelligibility in ambient noise levels of 120 db to the development of a system that coordinates unrelated hot-line interphone, strategic alert, and tie-in telephone subsystems into an integrated, flexible, command and technical communications network.

The paper discusses the selection of intermediate electrical levels, the provision for minimum changes in sound level regardless of the drastic changes in numbers of stations on any line, the design for AVC, and means for carrying overriding public-address-type announcements on all communications channels.

^{*} Sponsored by the Professional Groups on Space Electronics and Telemetry and Communications Systems. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

42.1. The Measurement of River Flow by the Use of Underwater Sound

G. E. MILLER, W. F. RICHARDSON, AND N. SEROTTA, Raytheon Co., Wayland, Mass.

A survey of presently known methods of measuring river flow is presented. The results of a study of the use of acoustic signals for the determination of flow velocity are given. Several different acoustic flowmeter systems are evaluated.

This work was supported by the U.S. Department of the Interior, Geological Survey, under Contracts 14-08-001-3993 and 14-08-001-5426.

42.2. Ultrasonic Flowmeter

H. DAHLKE AND W. WELKOWITZ. Gulton Industries, Inc., Metuchen, N. J.

An ultrasonic flowmeter has been developed which works on the principle of deflection of the sound beam by the liquid stream. An ultrasonic pulse is sent through a pipe transverse to the pipe axis. A receiver transducer which is on the opposite side of the pipe and is positioned on the side of the beam pattern measures the amplitude of the received signal. As the beam is shifted by the flow, this amplitude varies in proportion to the flow. Experimental results indicate that an eight-inch flowmeter can be built with about one per cent linearity for flow rates up to 3000 gallons per minute.

42.3. Optical Studies of Delay Line Transducers

RICHARD F. WEEKS, Richard D. Brew & Co., Inc., Concord, N. H.

Our present understanding of the properties of delay line transducers has been largely derived from electrical measurements of delay lines. Unfortunately, it is difficult to separate the effects due to the transducers from those due to the wave nature of the propagation of the signal media by purely electrical measurements. These difficulties can be overcome by measuring the acoustical signal from a transducer with a modified Schlieren system and looking at the transducer "from both ends." The photoelectric apparatus designed to make these measurements is described. It is shown that the ultrasonic field in fused silica may be described by the Fresnel and Fraunhofer diffraction formulas. An expression is derived for the insertion loss due to diffraction effects. Schlieren measurements of the ultrasonic fields are compared to the electrical admittance of several transducers to determine their coupling coefficient and dissipation as a function of frequence. Further studies planned at this laboratory are discussed briefly.

42.4. Ultrasonic Delay Line Analysis

D. L. SHILLING AND A. N. SILVER, Columbia University, New York, N. Y.

Based on earlier work by W. P. Mason and H. J. McSkimin, equivalent circuits are presented for the over-all transmission characteristics of an ultrasonic solid delay line. The resultant expressions for insertion loss, bandwidth, and ripple are plotted in terms of the normalized parameters of the piezoelectric transducer, the delay medium, and the electrical termination for 1) a backing material cemented to one face of the transducer, and 2) half-wavelength bonds (seals) cemented between the transducer and delay medium. These normalized curves constitute engineering design nomographs, which facilitate the simultaneous calculation of the insertion loss, bandwidth, and ripple, necessary to satisfy a particular over-all transmission characteristic. It is shown that employing backing materials a maximum 3-db bandwidth of 75 per cent may be achieved, while for a half-wavelength bond whose characteristic impedance approaches that of the delay medium, bandwidths of approximately 55 per cent are possible. In addition, experimental results are given for delay lines utilizing backing materials, and operating at a frequency of 10 mc. In this range, bandwidths of 65 per cent have been obtained. with an insertion loss of approximately 56 db.

42.5. A Comparison of Several Dispersive Ultrasonic Delay Lines Using Longitudinal and Shear Waves in Strips and Cylinders

ARTHUR H. FITCH, Bell Telephone Labs., Inc., Whippany, N. J.

Some recently developed ultrasonic delay lines employing the dispersion of elastic waves in solids possess advantages of cost and size in performing circuit functions heretofore accomplished with lumped parameter electrical networks. Several such delay lines are compared with regard to such features as compactness, versatility, delay-vs-frequency characteristics, and discrimination against unwanted signals.

42.6. Physical Principles and **Operational Characteristics** of Variable Ultrasonic **Delay** Lines

WALTHER ANDERSEN, Andersen Labs., Inc., West Hartford, Conn.

Variable acoustic delay lines typically consist of two pieces of fused quartz, each with an acoustic transducer, so designed that as one piece is moved with respect to the other, a variation in delay results. The mating surface of the two pieces of quartz must be very accurately ground, and a special viscous couplant is used between the two surfaces.

Ratio of delay variations as large as 50 to 1 have been obtained.

- Principal uses of these devices are:
- 1) In moving target simulators,
- In adjustable long delay lines suitable for 2) MTI or integrator applications,
- 3) In ECM devices,
- 4) As the control element of stable variable frequency oscillators.

42.7. New Techniques in Ultrasonic Delay Lines

DAVID L. ARENBERG, Arenberg Ultrasonic Laboratory, Inc., Jamaica Plain, Mass.

The first successful ultrasonic delay lines using multiple symmetry designs were based on regular polygons with a nonprime odd number of sides. Only a few facets were tilted from the regular position and the distance from the center, R_i , adjusted slightly.

To improve on the number of available designs, further investigations were made. Different criteria were used in selecting good designs, the most critical being the minimum aperture of the geometric beam at all the reflections. Maximizing this minimum results in a linear programming involving adjustment of the distances R_i .

Adjustment of all the angles allows an increase in the number of variables and better apertures. This last feature permitted the use of any polygon including prime and even numbers. Delay lines using these new designs have been made and found satisfactory.

SESSION 43*

Thurs.

10:00 A.M.-12:30 P.M.

Waldorf-Astoria Sert Room

EQUIPMENT AND SYSTEMS

Chairman: J. A. EGGERT, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.

43.1. Missile Master (AN/FSG-1)-System Functional Description

G. ROMANO, D. L. PRENTICE, AND I. HAYNE, The Martin Co., Orlando, Fla.

43.2. Missile Master (AN/FSG-1)-System Equipment Description

R. STASCHKE AND D. NODEN, The Martin Co., Orlando. Fla.

In these two papers, a description of the U. S. Army Missile Master, AN/FSG-1, antiaircraft defense system is presented. This description includes a project history and a review of Missile Master's part in the nation's over-all air defense system. Operational functions and equipment comprising the major subsystems, radar and tracking, tactical display, data communication, and fire unit integration, are described. These subsystems are discussed in terms of the data to be exchanged and the various displays and controls made available to the system's operators and the weapons commander. The integration of the Missile Master system with the U.S. Air Force semi-automatic ground environment (SAGE) and various missile systems, is also discussed.

^{*} Sponsored by the Professional Group on Mili-tary Electronics. To be published in Part 8 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

ciative network by a system of excitatory and inhibitory electrical connections. Patch panels and flexible controls permit a large number of elementary machine configurations to be studied.

40.4. A Magnetic Integrator for the Perceptron Program

J. K. HAWKINS, Aeronutronic Systems, Inc., Newport Beach, Calif.

A magnetic circuit possessing storage and output properties resembling those of W-unit memory elements in perceptron systems is described. Stored value is represented by net flux, and can grow or decay by application of pulses corresponding to activity of associated A-units. Readout is accomplished nondestructively by means of a field applied in a direction orthogonal to normal storage flux. The readout voltage is approximately proportional in both sign and amplitude to the net value of stored flux. Present circuits possess storage capacities of plus or minus 100 increments before saturating. Test results and integrator design considerations are discussed.

SESSION 41*

Thurs.

10:00 A.M.-12:30 P.M.

Waldorf-Astoria Astor Gallery

CIRCUIT THEORY: CURRENT CONTRIBUTIONS

Chairman: M. E. VAN VALKENBURG, University of Illinois, Urbana, Ill.

41.1. Transfer Function Synthesis of Active RC Networks

E. S. KUH, University of California, Berkeley, Calif.

A general method is found to synthesize voltage transfer functions with complex poles and zeros using RC elements and a practical active device (control source with finite input and output resistances). The active device can be a grounded base or a grounded collector transistor.

The method allows specified load resistance and often furnishes finite source resistance. The over-all network has a common ground and is economical in terms of the number of elements.

41.2. Broad-Band UHF Distributed Amplifiers Using Band-Pass Filter Techniques

FRED C. THOMPSON, *IIRB-Singer*, Inc., State College, Pa. The principle of distributed amplification has been applied to great advantage in the design of broad-band low-pass amplifiers. However, distributed amplifiers have generally been limited in application to the VHF region. This paper presents a technique for extending the useful operating frequency of distributed amplifiers to above 1000 mc. The technique basically employs planar triodes that are cascade-connected in a distributed band-pass amplifier. In addition, an active input network consisting of one or more grounded grid amplifier stages is used to optimize the amplifier sensitivity and frequency response.

A typical amplifier constructed according to the design theory presented above has a gain of 16 db ± 1 db over a frequency range from 500 to 1100 mc. Measured noise figure is less than 10 db.

41.3. A Fourier Series Time Domain Approximation

DOUGLAS R. ANDERSON, Hughes Research Labs., Culver City, Calif.

This paper presents a new method of approximating to the time domain function corresponding to any completely stable system function. Specifically, suppose we have a system function of the form

$$F(s) = \frac{p(s)}{q(s)}$$

where p(s) and q(s) are polynomials with positive coefficients whose zeros are only in the left half-plane and where

 $\deg (p(s)) < \deg (q(s)).$

Then by asymptotic formulas of Cerrillo one can compute a minimum positive number B such that the corresponding time function is uniformly small for $t \ge B$. And for t between 0 and any A the following trigometric series uniformly approximates the time function:

$$f_A(t) = \frac{\sqrt{2\pi}}{A} \left[\operatorname{Re}(F(0)) + 2\sum_{n=1}^{\infty} \operatorname{Re}\left(F\left(j\frac{n\pi}{A}\right)\right) \cos\left(\frac{n\pi t}{A}\right) \right].$$

Further, convergence of $f_A(t)$ is very fast whenever deg (q(s)) - deg (p(s)) is at least 4. Thus, in most cases of interest, this method requires fewer operations for the attainment of a given degree of approximations than do the standard methods of Guillemin, Truxal, and Floyd.

41.4. Spectral Measurements of Sliding Tones

W. GERSCH AND J. M. KENNEDY, School of Engrg., Columbia University, New York, N. Y.

The problem of specifying the value of the instantaneous frequency of a sliding tone at some instant in time is considered. A sliding tone is defined as a frequency which changes linearly with time. It is assumed that the specification is to be made by using a spectrum analyzer which is the equivalent of a bank of filters whose outputs are measured at a single point in time.

The problem of obtaining spectral measurements is treated by frequency domain considerations. Results are first developed for the spectral density of a finite fixed duration sinusoid of linearly time-varying frequency, from an interpretation of the Cornu spirals. This is referred to as the input spectrum. The response of a filter to a sliding tone is then computed by convolution of the input spectrum with the selectivity characteristics of the filter considered. The filter selectivity characteristic is obtained by Fourier transforming a periodic representation of the filter inpulsive response.

This approach to computation of the filtered characteristics of sliding tones illuminates an interpretation of the physically meaningful parameters of the filter and input and facilitates computation of the filtering action. In addition it helps suggest filter characteristics that are desirable from a spectral measurement point of view. Results are presented which indicate the response of a variety of filters to a sliding tone input.

The contribution of the approach presented to the evaluation of the response of filters to sliding tones has two major advantages over earlier efforts:

- Its usefulness for approximate analysis and the interpretation of the effects of filter and signal parameters,
- 2) The ease and economy of computation of accurate results.

For completeness, filter responses to a very rapidly sliding tone are discussed from the time domain point of view.

41.5. An Approach to the Synthesis of Linear Networks Through Use of Normal Coordinate Transformations Leading to More General Topological Configurations

E. A. GUHLEMIN, Elec. Engrg. Dept., Mass. Inst. Tech., Cambridge, Mass.

The proposed procedure presents a method of determining parameter matrices from given impedance functions through use of normal coordinate transformations, and realizes the pertinent network by identical tree configurations in single-element-kind networks having a general topological structure. Essentially the same procedure is applicable to passive bilateral networks and to nonpassive and/or nonbilateral ones, differing only in the appropriate normal coordinate transformation. Although tedious computations are involved, the availability of modern computers makes this method feasible and thus opens up a more general and potentially useful approach to network synthesis.

SESSION 42*

Thurs.

10:00 A.M.-12:30 P.M. Waldorf-Astoria

Jade Room

ULTRASONICS ENGI-NEERING—II

Chairman: WARREN P. MASON, Bell Telephone Labs., Murray IIill, N. J.

^{*} Sponsored by the Professional Group on Circuit Theory. To be published in Part 2 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

and $25\frac{1}{2}$ wavelengths in the transverse dimension are shown which provide excellent agreement with the theoretical curve.

Shown are curves of theoretical ripple, as a function of the number of terms of the Fourier series used, for antennas whose main lobes are pointed at 3° and 15° below the horizon. The curves show that it is theoretically possible to synthesize a $\csc^2 \theta$ pattern to a tolerance of 0.4 db with an aperture efficiency corresponding to HPBW = 90/D/ λ as compared with a typical 120/D/ λ for a reflector-type aperture.

38.4. Determination of Optimum Primary Feed Ellipticity Setting to Obtain Circular Polarization from Reflector-Type Antennas

L. J. KUSKOWSKI, Airborne Instruments Lab., Melville, N. Y., AND A. M. McCoy, Raytheon Manufacturing Co., Wayland, Mass.

A problem prevalent in the use of circular polarization feeds for large parabolic or shaped reflectors is the determination of the optimum primary feed peak ellipticity to obtain circular polarization on the peak of the secondary beam. This problem can usually be solved by a "cut and try" method providing that an antenna range which is long enough and free of reflections is available.

However, with very large reflectors, this can become a serious problem. A simpler approach has been developed which requires neither the "cut and try" method nor the considerable range. This method does not require the use of the reflector and is not limited by reflector size.

Experimental results obtained with an X-Band Scale Model Reflector have verified that the above technique is valid and gives excellent results.

SESSION 39*

Wed.

2:30-5:00 P.M.

Coliseum Morse Hall

MICROWAVE INTERACTION WITH MATTER

Chairman: WILLIAM W. MUMFORD, Bell Telephone Labs., Whippany, N. J.

Panel Members; S. C. BROWN, Mass.
Inst. Tech., Cambridge, Mass.; C. L.
HOGAN, Motorola Semiconductor
Div., Phoenix, Ariz.; AND
H. KROEMER, Varian Associates, Palo Alto, Calif.

39.1. Recent Progress in Microwave Beam, Plasma, and Solid-State Devices

LESTER M. FIELD, Hughes Aircraft Co., Culver City, Calif.

A comparison will be made in many frequency ranges between electron beam, gas beam, plasma and solid-state devices used for microwave amplification and oscillation. Parametric, maser, and traveling-wave principles, among others, will be described using these media. Optical masers and plasma sources of microwave energy will be reviewed.

Through improvements in electron beam focusing techniques using magnetic or electric periodic focusing and new forms of microwave circuits, unprecedented levels of gain and trequency have been obtained recently.

Comparable improvements in bandwidth, noise figure, temperature of operation, and frequency using the other media will be discussed.

39.2. Microwave Interaction with Plasmas

R. G. BUSER AND P. WOLFERT, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.

The interactions of microwaves and plasmas in both the presence and absence of magnetic fields and the theoretical principles involved are briefly developed. An experimental apparatus was built wherein high-energy plasmas can be produced in a magnetic field for about 50 μ sec.

A microwave beam, emitted from a conventional horn antenna, impinges on the plasma. The energy of the beam is divided into four parts: transmitted, reflected, absorbed, and scattered energy. An experimental setup is described that measures the transmitted, reflected, and scattered energies during the production and recombination periods. These measurements give important information about density, collision frequency, temperature, and plasma movement. Several problems of interpretation of these measurements are discussed.

39.3. A New Semiconductor Microwave Modulator

H. JACOBS, F. A. BRAND, AND M.
BENANTI, U. S. Army Signal R & D
Lab., Fort Monmouth, N. J., AND
R. BENJAMIN, Monmouth College, West Long Branch, N. J.

Experiments have been conducted in which semiconductor rods of germanium are inserted in a waveguide parallel to the direction of the electric field. Upon exposure to light or the injection of minority current carriers by means of a p-n junction, the conductivity of the semiconductor is changed. The changes in conductivity, in turn, cause variations in the absorption of microwave energy. This effect has been designed into a device which offers a possibility of microwave amplitude modulation. Performance data and design information relating to the role of the semiconductor lifetime, the effects of various types of trapping centers, and other electrical properties will be described.

SESSION 40*

Thurs. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Starlight Roof

ADAPTIVE NETWORKS

Chairman: MARVIN L. MINSKY, M.I.T. Lincoln Lab., Lexington, Mass.

40.1. Pattern Recognition with an Adaptive Network

LAWRENCE ROBERTS, M.I.T., Lincoln Lab., Lexington, Mass.

Adaptive networks produce an output by means of interconnections and weighted values derived from previous experience. Several adaptive networks have been designed to test practicality of simple noninterconnected networks for use in pattern recognition. The particular networks which will be described are capable of recognizing many patterns or characters with a high degree of probability. These networks use a single layer of neurontype elements which provide a type of correlation function on the input matrix—a technique which has proved successful in many tests.

40.2. On Predicting Perceptron Performance

R. DAVID JOSEPH, Cornell Aeronautical Lab., Inc., Buffalo, N. Y.

Perceptrons are devices intended to simulate a portion of the logic of the brain concerned with perception. The memory of a perceptron is contained in the values of many components. Perceptrons are classified according to the manner in which these values are modified as result of contacts with the environment. Mathematical analyses are available for the three main types of perceptrons which have a logical depth of two. The capabilities of these systems to classify their environment will be presented, as well as those results for systems of greater logical depth that are available.

40.3. The Mark 1 Perceptron-Design and Performance

J. C. HAY, F. C. MARTIN, AND C. W. WIGHTMAN, Cornell Aeronautical Lab., Inc., Buffalo, N. Y.

An experimental model of a perceptron has been constructed and is being used to verify earlier mathematical predictions for its type, and to compare its cognitive abilities to those of biological systems. Visual input is provided through a lens and a retina of 400 photocells. Although the machine is too slow for rapidly changing patterns such as real-time speech, sensory modalities other than visual are also possible. Stimulus patterns received by the sensory units undergo a scattered and otherwise complex projection onto the inner, asso-

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^{*} Sponsored by the Professional Group on Electronic Computers. To be published in Part 2 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

the receiver. In this respect the PM system is similar to FM communication systems which utilize an "FM improvement factor." Several examples are presented to indicate the magnitude of the PM improvement factor when the signal is a sine wave variable throughout the audio band, and a novel type of closed-loop phase detector is suggested which is linear over a very wide range of phase angles.

37.2. An Improved Decision Technique for Frequency Shift Communications Systems

ELMER THOMAS, Page Communications Engineers, Inc., Washington, D. C.

An improved mark-space decision technique for frequency shift systems is discussed. Fading has been observed to be frequency-selective on ionospheric-scatter and high-frequency communications circuits. Conventional decision techniques are shown to result in a high error liability when deep fades occur on either the mark or space frequency. A decision circuit is described, which effects an improvement approximately equal to an extra order of diversity where fading in mark and space frequencies is independent. A theoretical analysis showing the degree of improvement in signal detectability offered by this new technique is given. Experimental data obtained on operating circuits, as well as in laboratory tests, are presented, verifying the theory and demonstrating the practicability of the technique.

37.3. High Sensitive Receiving Systems for Frequency Modulated Wave

M. MORITA AND S. ITO, Nippon Electric Co., Ltd., Kawasaki-shi, Kanagawa-ken, Japan.

This paper presents some methods of improving the sensitivity in an FM receiver. The conventional FM system has a defect called the threshold level. We have improved this threshold level to a great extent by several methods, and reliable communication has become possible even at weak electromagnetic field intensity.

One of these methods is to demodulate the received FM signal after it is combined with a large sinusoidal voltage, and to apply the FM negative feedback technique. Another method is to detect the phase difference between the received FM or PM signal and a large sinusoidal voltage, and also to apply the FM negative feedback technique.

37.4. An Improved Multiplex Voice Frequency Carrier System

BERNARD TENNENT, Philips Electronics Industries Ltd., Toronto, Ont., Canada.

Using newly developed ferrite tone channel filters, discriminators, and oscillators, multiplexing with channels spaced only 85 cps apart has been made possible in a \pm 30-cps shift system—thereby almost doubling the usual number of available channels in an allotted tone spectrum.

At least 40-db interchannel attenuation is retained, while the 60-cps wide acceptance band is practically flat, symmetrical and stable. In another application, the above tone units in conjunction with a specially designed steep band stop filter, enable superposition on a voice carrying wire line the facilities of up to 4 telegraph duplex channels. There is a small degradation of the speech intelligence, as only a narrow portion of the voice spectrum is sacrificed.

The audio networks are built with LC-type resonators, in which ()'s of up to 1000 have been achieved with recently available high permeability ferrite cores.

A complete operating system is described.

37.5. Model of Impulsive Noise for Data Transmission

PIERRE MERTZ, Long Beach, N. Y.

It has often been found more necessary in the engineering of data transmission systems to consider impulsive noise than conventional Gaussian "white" noise. A model is proposed for the impulsive noise, which describes empirically an amplitude distribution and a time distribution. Because the latter has in experimental work been described principally in terms of error occurrences, the description is translated into these. The notable characteristics of impulsive noise are that at low occurrence frequencies the amplitudes are much larger than for Gaussian noise, and that impulses or errors tend to be more "bunched" than expected from a Poisson distribution.

SESSION 38*

Wed.

2:30-5:00 P.M.

Coliseum Marconi Hall

ANTENNA PATTERN SYNTHESIS

Chairman: ALLEN S. DUNBAR, Lockheed Missile Systems Div., Sunnyvale, Calif.

Panel Members: R. C. SPENCER, The Martin Co., Baltimore Md.; P. A. BRICOUT, Emerson Res. Labs., Silver Spring, Md.; AND R. BICK-MORE, Hughes Aircraft Co., Culver City, Calif.

38.1. Derivative Control in Shaping Antenna Patterns

A. KSIENSKI, Hughes Aircraft Co., Culver City, Calif.

An antenna pattern synthesis method is presented for arrays of fixed element spacing that permits one to approximate desired patterns with arbitrarily prescribed error criteria. Thus either smoothness, linearity, or curvature may be specified for certain parts of the pattern. If desired, an equal ripple approximation may also be very closely approached. The method is seminumerical in nature and may involve several perturbations depending on how closely one desires to approach the optimum. Several numerical examples are worked out in detail, and an experimental verification is also provided.

38.2. Some New Methods of Analysis and Synthesis of Near-Zone Fields

MING-KUEI IIU, Elec. Engrg. Dept., Syracuse University, Syracuse, N. Y.

In this paper, the following new methods of analysis and synthesis of near-zone, including Fresnel-region, fields will be presented:

- A principle of subdivision and methods of near-zone field analysis. This principle makes possible the use of far-field techniques in near-zone field analyses.
- A near-zone field synthesis method based upon the principle of subdivision. This method bears a close resemblance to the Woodward method of far-field synthesis.
- 3) A focusing theorem and a second nearzone field synthesis method. This method transforms a near-zone synthesis problem into a far-field problem, therefore any synthesis method applicable in the far-field can also be applied here.
- 4) Fresnel-region field analysis of a circular aperture. The results are expressed in terms of a class of new functions which are closely related to the Lommel's functions of two variables. Simple method of evaluating such functions will also be given.
- A second focusing theorem and Fresnelregion field synthesis. The synthesis is also based upon far-field methods.

Examples of both near-zone synthesis methods will be included. Merits of each method will also be discussed.

38.3. Synthesis of CSC² θ-Type Antenna Patterns Using Two-Dimensional Surface Wave Arrays

H. W. COOPER AND H. R. MCCOMAS, Westinghouse Electric Corp., Baltimore, Md.

The development and design principles for a unique flat array (0.032 inch thick at Ka band) of nonresonant slot radiators are described. A surface wave of the dielectric image line type is used as a transmission line to excite a transverse array of slot radiators. The procedure for designing an array of this type is outlined, and it is shown that a dielectric image line of the appropriate size has a field in the transverse direction that decays in almost identically the same fashion as the Fourier coefficients of typical shaped beam patterns used for ground mapping. Thus, an array can be constructed which uses identical slot elements in the transverse plane to achieve the proper amplitudes of excitation coefficient. These radiating slot elements are displaced in the longitudinal plane to achieve the required phase coefficient.

Measured radiation patterns of an array 425 wavelengths in the longitudinal dimension

^{*} Sponsored by the Professional Group on Antennas and Propagation. To be published in Part 1 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

35.3. The Transient Effect in Capacitor Leakage Resistance Measurements

RAYMOND W. FRANCE, Hughes Aircraft Co., Culver City, Calif.

The leakage resistance of capacitors as a function of time is a characteristic to be considered in the choice and control of capacitors, especially in many modern military electronics applications. This paper, by the use of transient circuit analysis, shows why most leakage resistance data accumulated in the electronics industry for capacitors are invalid. Valid data can be obtained by using low resistance meters to make such measurements. Valid experimental curves of leakage resistance vs time for thirteen types of capacitors, each type employing a different dielectric combination, are slown; the significance of these curves is discussed.

35.4. Dynamic Temperature Coefficient of Microelement Inductors

G. HACSER, RC.1, Somerville, N. J.

The application of ferrite toroids in frequency-dependent circuits requires specification of the inductance stability. In particular, the temperature characteristic of the ferrite becomes an important circuit parameter. Conventional point-by-point measurements of inductance-temperature characteristic do not always reveal deficiencies of the ferrite material. Therefore, a dynamic test procedure was designed to allow an extensive study of the ferrite temperature coefficient. The test equipment produces a continuous chart of the inductance as a function of temperature. The same equipment is also capable of providing graphs of inductance as a function of ac and de drive, as well as the temperature coefficient of capacitors and tuned circuits. The National Bureau of Standards reference core proved that the dynamic-temperature-coefficient equipment has sufficient accuracy.

35.5. A New Automatic Method for the Design of Low-Voltage Transformers on the IBM 704

DAVID A. FRANKS, Westinghouse Electric Corp., Baltimore, Md.

A new technique for designing low-voltage transformers on a high-speed digital computer has been developed and programmed for the IBM 704. The program developed provides for the design of transformers having the following characteristics: 1) N secondary windings, 2) up to N shields, any of which may be wire-wound or foil, and 3) up to three voltage taps on each winding. Cases where N is as large as 10 can be handled successfully. The program produces the manufacturing specifications in a form suitable for reproduction for use by shop personnel in manufacturing the transformer. The paper describes the technique and some of the results obtained from the computer program using the technique.

SESSION 36*

Wed.

2:30-5:00 P.M.

Waldorf-Astoria Sert Room

STEREOPHONIC SOUND REPRODUCTION

Chairman: BENJAMIN B. BAUER, CBS Labs., Stamford, Conn.

36.1. Stereophonic Sound Reproduction

HARRY F. OLSON, RCA Labs., Princeton, N. J.

This paper presents the following aspects of stereophonic sound reproduction: the fundamental conditions of frequency range, volume range, reverberation and spatial sound patterns required to obtain realistic stereophonic sound reproduction; the reproduction of stereophonic sound in rooms and automobiles; and stereophonic sound systems in the consumer complex.

36.2. Psychoacoustics of Stereophonic Reproduction

R. L. HANSON, Bell Telephone Labs., Inc., Murray Hill, N. J.

When one listens to a transient-type signal from two equidistant loudspeakers in free space, he experiences the sensation of a single well-defined source. The apparent position of the source can be altered by adjusting the relative outputs of the two speakers, by delaying the signal from one, or by a combination of level and delay adjustments. If the transienttype signal is replaced by a steady-state single frequency tone, the virtual source is again well-defined and can be changed in azimuth position by altering the speaker outputs as long as the two sources are in phase. However, as one source is delayed, the virtual source becomes indistinct and localization becomes difficult. Tests will be described which indicate that when one listens to a steady-state single frequency tone, as from a pair of sources in free space or from a single source and its images in a room, he experiences the sensation of a welldefined single source only when the resultant time and pressure differentials correspond to those which could result from a single source in free space. This indicates that the apparent position of a steady-state single frequency source in a room bears a rather complex relation to the position of the actual source. Correct localization is dependent on transients in the signal and may be influenced by visual cues.

36.3. Some Considerations in Design and Application of a Compatible Magnetic Tape Cartridge

MARVIN CAMRAS, Armour Res. Foundation, Illinois Inst. Tech., Chicago, Ill. Magnetic tape may be offered in cartridge form at a price competitive with phonograph disks. Cartridges of different sizes are designed either for superior quality or for maximum tape economy. All of these will operate on present-day machines, as well as on automatic designs. A cartridge-changer allows records to be played in sequence. The erase feature offers interesting possibilities for sale of pure music, separate from the sale of cartridges.

36.4. A 1-7/8 IPS Magnetic Recording System for Stereophonic Music

P. C. GOLDMARK, C. D. MEE, AND W. P. GUCKENBURG, CBS Labs., Stamford, Conn.

A new magnetic recording and reproducing system has been developed leading to a prerecorded cartridge substantially smaller than other known media. New approaches to the recording and reproducing system, as well as new components, are described.

36.5. Automated Magnetic Tape Cartridge Mechanisms

JOHN D. GOODELL, CBS Labs., Stamford, Conn.

Problems relating to the geometry of magnetic tape cartridges are discussed, along with the relative advantages of various solutions. Designs for tape transports and cartridge changer mechanisms are described, together with systems for fully automated programming of all operations.

SESSION 37*

Wed.

2:30-5:00 P.M.

Coliseum Faraday Hall

COMMUNICATION SYSTEM TECHNIQUES

Chairman: ANATOLE MINC, Tele-Signal Corp., Glen Cove, N. Y.

37.1. Analysis of a Phase Modulation Communications System

ROBERT L. CHOATE, Jet Propulsion Lab., California Inst. Tech., Pasadena, Calif.

A communication system is analyzed which utilizes phase modulation (PM) and phase detection. The analysis shows that when the total RF power level is significantly above the receiver threshold, a margin of safety exists which may be traded for additional signal-tonoise ratio (SNR) at the demodulated output of

^{*} Sponsored by the Professional Group on Audio. To be published in Part 7 of the 1960 IRE International Convention Record.

^{*} Sponsored by the Professional Group on Communications Systems. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

W. C. FRY, Westinghouse Electric Corp., Pittsburgh, Pa.

A single n-p-n-p switch rectifier or Trinistor device is used in an inverter circuit of unique design to provide simultaneously the 20-kc ultrasonic power and the dc polarization to a spaced lamination magnetostrictive transducer. The inverter utilizes feedback which makes the circuit self-oscillating. This feedback also provides an automatic frequency tracking feature which causes the inverter to be stable at or near the resonant frequency of the transducer and to follow resonant frequency variations caused by changes in transducer loading. Included in this paper are comments on the Trinistor device and its operating characteristics in high frequency service and a detailed discussion of the particular circuit used in the ultrasonic power source.

34.4. Ultrasonic Cleaning Tests for a Variety of Driving Waveforms

R. C. HEIM, Westinghouse Electric Corp., Pittsburgh, Pa.

In order to determine the relative effectiveness of pulsed and continuous driving waveforms in ultrasonic cleaning systems, a series of cleaning tests was carried out using a soil preparation of aluminum oxide and grease similar to that described by Koontz and Amron in ASTM Special Technical Publication No. 246. These tests show that a pulsed waveform provides better cleaning action only when the average RF power is below two or three watts per square inch of transducer radiating area. Above this level of driving intensity the unmodulated or continuous wave input provides better ultrasonic cleaning for a given exposure time. Information on the sensitivity of the experiment to soil thickness, detergent concentration, and condition of the cleaning solution is included.

34.5. The Effectiveness of Ultrasonic Degreasing as Measured by Radiotracer Techniques

E. L. ROMERO AND H. A. STERN, RCA, Lancaster, Pa.

The ability of ultrasonic cleaning to remove radioactive grease from a television-picturetube cathode was studied by means of a technique in which radioactive stearic acid was applied to the inside of deep-drawn cathode cups. The degree of soil removal was measured after the labeled parts were cleaned in a detergent solution by means of ultrasonic cavitation. Effectiveness of degreasing was studied as a function of the following factors: 1) the addition of ultrasonic energy, 2) the load of parts being cleaned, 3) the "de-aeration" of the load being cleaned, 4) the size of the ultrasonic equipment, 5) the frequency of the ultrasonic equipment, and 6) the input power of the ultrasonic equipment. As a result of this study, a meaningful test and test apparatus were designed for the measurement of the degreasing potential of any ultrasonic cleaning device.

34.6. A Spaced Lamination Transducer for Industrial Use

E. B. WRIGHT, Westinghouse Electric Corp., Pittsburgh, Pa.

The spaced lamination transducer, designed primarily for the ultrasonic irradiation of cleaning and electroplating baths, is discussed in detail. Through the use of the spaced lamination technique, good acoustic loading of any magnetostrictive material can be obtained, thereby making it possible to utilize such materials more effectively in an industrial transducer unit. Comments on a variety of active materials and other design variables are included along with experimental data on the spaced lamination device. Final evaluation of relative cleaning effectiveness is made using a specially prepared aluminum oxide and grease soil similar to that described by Koontz and Amron of Bell Laboratories in ASTM Special Technical Publication No. 246.

34.7. An Efficient Low-Cost Ultrasonic Transducer for Use in Remote Control and Carrier Frequency Applications

FRANK MASSA, Massa Div., Cohu Electronics, Inc., Hingham, Mass.

A small, efficient ultrasonic transducer will be described which incorporates a novel vibrating system that includes a specially electroded piezoelectric disk bonded to a second plate. Two basic designs will be discussed, one using the external surface of a rugged, waterproof housing as the radiating surface, and a second which makes use of the resonant mode of a freely suspended vibrating disk. The transducer design permits operation in air at frequencies in the range 15 to 60 kc. At the higher ultrasonic frequencies bandwidths up to about 8 kc are possible, which permits use of the transducer in portable "walkie-talkie systems using an ultrasonic carrier to replace the radio frequency transmitter.

SESSION 35*

Wed.

2:30-5:00 P.M.

Waldorf-Astoria Jade Room

COMPONENT PARTS

Chairman: ROBERT ASHBY, Autonetics, Inc., Downey, Calif.

* Sponsored by the Professional Group on Component Parts. To be published in Part 6 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

35.1. Magnetostrictive Ultrasonic Delay Lines for a PCM Communication System

D. AARONSON AND D. B. JAMES, Bell Telephone Labs., Murray Hill, N. J.

A servo operated delay line pad and a temperature compensited delay line memory, both magnetostrictively driven at 1.5 mc, have been used in an experimental PCM communications system. (II. E. Vaughan, "Research Model for Time-Separation Integrated Communication," *Bell Sys. Tech. J.*, vol. 38, p. 909; July, 1959.)

The delay line pad automatically compensates for external delay changes as small as plus or minus 8 m/sec at a rate of 75 m/sec per second. The delay line memory stores 192 hits which are available both serially or in parallel with an access time of 125 μ sec. Both applications use the same basic delay lines which consist of a length of 0.003-inch diameter supermendur wire, two tiny solenoids, and a support.

35.2. The Reliable Application of Electronic Component Parts

H. LANE DUDLEY, Melpar, Inc., Alexandria, Va.

Both the selection and the proper use of component parts are the keystones of welldesigned electronic equipment. In this paper it will be pointed out that the selection of reliable components, taking fullest advantage of military approved parts, is only the first part of the job. The second part must be handled by engineers familiar with component parts and basic failure rate data—the engineering reliability group.

It must be recognized that most design engineers are not component application specialists and that there are many pitfalls encountered in proper usage. In a field survey of 18,000 avionics equipments conducted by Hughes Aircraft Company and reported in *Avialion Week*, September 30, 1957, it was stated that of 5 million parts installed, 5 per cent were responsible for 55 per cent of all part failures.

To minimize this misapplication problem, the circuit designer should be required to check the performance of the breadboard over the range of temperatures at which it must operate. After the circuit has passed breadboard tests, a basic and derated failure rate should be assigned the circuit by the reliability group. At this time the reliability group automatically checks circuit stresses on components such as peak inverse voltages, dissipation, peak currents, etc. After discrepancies have been ironed out with the design engineers the prototype design is frozen. However, the reliability group should continue to monitor the use of components. After the prototype is built there are many hours of debugging and test. During this period the reliability group should receive a report on every failed part and the type of failure-shorted, leaky capacitor, open, etc. These parts must be identified as to actual circuit placement as well as type, thereby enabling the reliability group to further pinpoint problem areas.

The component failures may often be eliminated by the selection of a component with a higher rating, and in extreme cases circuit redesign is necessary. It is by use of the methods outlined above that a more reliable equipment can be developed through proper choice and use of reliable components. quency signal propagating in the waveguide. The calorimeter is designed for high-power operation and for absorption of signal(s) of any direction or waveguide mode. Thus, measurement of spurious outputs of high-power tubes can now be made quickly and accurately without recourse to mathematical analysis previously required.

SESSION 33*

Wed.

Waldorf-Astoria

2:30-5:00 P.M.

ELECTRONIC COMPUTERS AND CIRCUIT THEORY: HOW EACH TECHNOLOGY CAN HELP THE OTHER

Starlight Roof

Chairman: JEAN H. FELKER, A T&T Co., New York, N. Y.

33.1. Switching and Memory Criterion in Transition Flip-Flops

D. O. PEDERSON AND D. K. LYNN, University of California, Berkeley, Calif.

It is well known that the magnitude and length of the minimum input trigger pulse is closely related to the amount of memory of the flip-flop. A study of this relationship is made in this paper. Any energy or charge storage element or mechanism cannot serve as the flip-flop memory. A criterion in terms of a simple physical argument has been established to determine which storage elements provide the necessary memory function. The minimum amount of memory is determined from an analysis of the initial conditions of the regenerative switching action. Piece-wise linear analysis techniques are used starting from either conventional circuit analysis or from a natural mode analysis. Examples of capacitivity-coupled and inductivity-coupled circuits, including experimental confirmation, are used as illustrations.

33.2. Monte Carlo Analysis of Transistor-Resistor Logic Circuits

Y. C. HO AND W. J. DUNNET, Sylvania Electronic Systems, Necdham, Mass.

This paper describes a general approach to statistical investigation of properties of complex transistor switching circuits. In particular, we are concerned with the TRL circuits which use resistive coupling between grounded emitter stages to perform the logical NOR function. An important consideration in the design of TRL systems is the propagation delay of pulse signals through various levels of these circuits. A mathematical model of the delay has been constructed which is a complicated function of the circuit variables as well as of the intrinsic parameters of the transistors involved. A computer program was written to simulate the model on an IBM 709. By using measured statistical data of transistor parameters and randomly sampled circuit variables as input, a Monte Carlo analysis of the distribution of propagation delay is carried out. The results were evaluated by means of nonparametric statistics and verified by actual experimental measurements.

33.3. An Analog Computer Nyquist Plotter

E. A. GOLDBERG, Space Technology Labs., Inc., Los Angeles, Calif.

In most analog computer simulations of automatic control systems, the simulation is performed in the time domain. However, for stability analysis and system design it is very useful to be able to obtain frequency domain information from the same simulation. This paper describes an automatic Nyquist plotter which can be used for this purpose.

The operation of this Nyquist plotter is based upon a sampling technique. The information is obtained directly in Nyquist form and can be presented on an x-y plotter. The device has been applied to linear (continuous and sampled-data) control systems. With no special techniques, an accuracy of 2 to 3 per cent was easily achieved.

33.4. Smoothing and Prediction of Time Series by Cascaded Simple Averages

R. B. BLACKMAN, Bell Telephone Labs., Inc., Murray Hill, N. J.

Theories of optimum smoothing and prediction of time series have been worked out under a variety of assumptions regarding the characteristics of the data. The results of these theories are important insofar as they establish a ceiling and a standard of comparison for actual schemes. In practice, however, attempts to achieve optimum results usually require large numbers of memory slots and large amounts of computation. Several approximate schemes have been worked out to reduce one or the other or both of these practical requirements. The scheme described in this paper is one which was devised to reduce the amount of computation required.

33.5. Synthesizing Minimal Stroke and Dagger Functions

JOHN EARLE, IBM Corp., Poughkeepsie, N. Y.

The techniques of the functions tables, theorems of Boolean algebra, and Karnaugh maps are extended to provide synthesizing and minimizing methods for the stroke and dagger functions. This paper describes the transformations among these and more familiar functions

These methods are applicable to NOR circuits and all AND-Inverter, OR-Inverter type circuits performing these functions.

The work in this paper was completed at the Underwood Laboratories, Hartford, Conn., August, 1958, prior to the author's association with 1BM. Wed.

2:30-5:00 P.M.

Waldorf-Astoria Astor Gallery

ULTRASONICS ENGI-NEERING-I

Chairman: WILLIAM P. RANEY, Harvard University, Cambridge, Mass.

34.1. Eigen Coupling Factors and Principal Components, The Thermodynamic Invariants of Piezoelectricity

H. G. BAERWALD, Sandia Corp., Albuquerque, N. Mex.

Piezoelectricity is represented in thermodynamically invariant description as a phenomenon relating to two distinct forms of reversible energy density present simultaneously. This is largely irrelevant to, and basically simpler than, the customary tensorial description. Each piezoelectric crystallographic class is in general associated with three eigen coupling factors and associated principal strain-stress and fielddisplacement component sets, representing stationary values of the ratio: elasto-dielectric to self-elastic-dielectric energy density. Depending on crystallographic symmetry, some eigen coupling factors may be zero or equal to each other. The absolutely largest factor is characteristic of the optimum capability of the medium to transduce electromechanical power.

34.2. Piezomagnetic Ceramic Transducers

OSKAR E. MATTIAT, Curtiss-Wright Corp., Santa Barbara, Calif.

The mechanical, magnetic, and magnetomechanical properties of a new piezomagnetic ceramic material are described and compared with presently used magnetostrictive material. Piezomagnetic ceramic transducers have negligible eddy current losses and, therefore, a much better magnetoacoustic efficiency than the conventional magnetostrictive transducers.

A comparison with piezoelectric ceramics (barium titanate PZT, etc.) shows that piezomagnetic ceramics exhibit a lower coupling coefficient, yet a higher Curie point temperature. The new material is well suited for high acoustic intensity transducers, and has great advantages for applications at elevated temperatures. Certain compositions of piezomagnetic ceramics have a zero temperature coefficient and a very high mechanical quality factor (up to 5000) which make them excellently suited for resonator elements in electric wave filters.

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^{*} Sponsored by the Professional Group on Ultrasonica Engineering. To be published in Part 6 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

28.3. Conceptual Design of a General-Purpose Telemetry System

W. F. LINK, Aeronutronic Div., Ford Motor Co., Newport Beach, Calif.

A broad and general study of the telemetry field, particularly problem areas and means for improvement, has been concluded. This program was tri-service sponsored and had as a primary objective a comparison of modulation methods to find the basic system best suited to handle the bulk of user requirements at the primary test ranges. It was found that most modulation methods which have been used offer one or more salient advantages for a specific application, but a new system is proposed which will be optimum or nearly so for the great majority of user needs. Specifications of parameters and experimentally determined performance of the conceptual system are given.

28.4. Detection Levels and Error Rates in PCM Telemetry Systems

A. V. BALAKRISHNAN AND I. J. ABRAMS, Space Technology Labs., Inc., Los Angeles, Calif.

Space and missile telemetry systems employing PCM-FM and PCM-PM are analyzed to determine optimal detection criteria and associated error rates as a function of relevant system parameters. Results are presented in both analytical and graphical form to aid in system optimization. A feature of the analysis is that the usual admittedly inaccurate assumption that the noise is additive and Gaussian in the decision circuitry has been avoided. In particular, exact first-order statistics of the demodulated FM and PM outputs are determined here for the first time. The implications of the results to the communication efficiencies of these systems are also discussed.

28.5. A Highly Precise FM/FM Telemetering Device

Howard K. Schoenwetter, *Hoover Electronics Co., Timonium, Md.*

Vernitel is a precision device which makes possible the transmission of data with high accuracy over standard existing FM/FM telemetering systems.

The heart of the Vernitel is a special quantizer which converts the input signal voltage into two voltage components—a quantized voltage and a vernier voltage—whose sum is equal to the input voltage. After quantization the voltage components are amplified and fed to two standard FM subcarrier oscillators.

At the receiving station the corresponding voltage components are obtained from two standard FM discriminators and recombined according to simple rules to form the original input signal.

The Vernitel was designed to operate under severe missile and aircraft environment and will normally result in an accuracy improvement of 8 to 10 times that obtainable with standard FM/FM equipment.

SESSION 29*

Wed. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Grand Ballroom

SEMINAR ON 1959 ITU GENEVA CONFERENCE

Master of Ceremonies: F. C. DE WOLF, Dept. of State, Washington, D. C.

Chairman: COMMISSIONER T. A. M. CRAVEN, FCC, Washington, D. C.

Panel Members: COMMISSIONER R. HYDE, FCC, Washington, D. C.; E. A. ALLEN, FCC, Washington, D. C.; G. G. GROSS, Secretary General, ITU; A. L. LEBEL, Dept. of State, Washington, D. C.

From August, 1959 to January, 1960, a world-wide radio conference was held in Geneva under the aegis of the International Telecommunication Union to review and revise the regulations governing the international allocation and utilization of the radio spectrum. The inside story of this important conference will be presented by a panel of leading participants.

SESSION 30**

Wed.

10:00 A.M.-12:30 P.M.

Coliseum Faraday Hall

COMMUNICATION SYSTEMS DESIGN

Chairman: EUGENE D. BECKEN, RCA Communications, Inc., New York, N. Y.

30.1. Equipment Configuration and Performance Criteria for Fully Optimized Tropospheric-Scatter Systems

CHARLES A. PARRY, Page Communications Engineers, Inc., Washington, D. C.

The optimum system is examined with the aid of a basic equation related to channel signal-to-noise ratio. It is shown on this basis that

SESSION 28*

Wed. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Sert Room

SPACE TELEMETRY

Chairman: DONALD M. CULLER, ITT Labs., Fort Wayne, Ind.

28.1. A Versatile Data Processing Facility

J. P. RANDOLPH, Johns Hopkins University, Silver Spring, Md.

A highly versatile data processing facility has been established at the Applied Physics Laboratory, Johns Hopkins University, for use in the Navy Polaris, Terrier, Tartar, and Talos missile programs.

The data processing equipment accepts as inputs PAM, PDM, PCM, and FM-FM telemetry information, as well as certain forms of nontelemetered digital systems data. These data are processed as necessary for graphical display, analog computer, or entry into a Remington Rand 1103A computer.

Semiautomatic digitizers are used to translate film and strip-chart data to IBM cards for entry into an IBM 650 computer. Two X-Y plotters operate automatically from IBM cards.

28.2. Evaluation of Modulation Methods for Telemetry Usage

M. RUDIN AND D. CHILDERS, Aeronutronic Div., Ford Motor Co., Newport Beach, Calif.

Methods of telemetry systems evaluation and comparison are presented from the standpoint of operational requirements and information theory, and related to the results of laboratory experiments. Optimization of each of four systems of particular interest (PAM-FM, PDM-FM, PCM-FM, FM-FM) is described briefly. It was found in some cases that the systems as commonly adjusted are not optimum and that significant performance improvement is attainable.

Comparative merits of the four systems are presented for varying accuracy by means of optimized performance charts, and the crossover between analog and digital systems is demonstrated and explained qualitatively. The theoretical efficiency of a hybrid, variable accuracy system is discussed on the basis of maximum spectrum utilization and minimum requirement on received power, consistent with the ability to accommodate a wide range of user requirements.

^{*} Sponsored by the Professional Groups on Antennas and Propagation, Broadcasting, Communications Systems, Space Electronics and Telemetry, and Vehicular Communications. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION REC-ORD.

ORD. ** Sponsored by the Professional Group on Communications Systems. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

^{*} Sponsored by the Professional Group on Space Electronics and Telemetry. To be published in Part 5 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

27.2. Tomorrow's Technology-

Functional Electronic Blocks

COL. WILLIAM S. HEAVNER,

Wright-Patterson AFB, Ohio.

The feasibility of functional electronic

closed servo loop. Squelch circuits are unnecessary because of the absence of a carrier oscillator.

Problems connected with the matrixing of the L and R signals as well as with RF, IF bandwidths are reviewed.

26.6. A New Concept in Transistor Converters

L. PLUS AND R. A. SANTILLI, RCA, Somerville, N. J.

This paper describes a method of obtaining an improved AGC characteristic in transistorized broadcast-band receivers. This improvement is made possible through the use of a new device developed specifically for this application. This device may be used in any transistorized broadcast band receiver, but it is particularly advantageous when an RF stage is not used. Several typical circuits using this new device are presented, and their AGC characteristics are compared with the AGC obtainable with conventional techniques.

SESSION 27*

Wed.

10:00 A.M.-12:30 P.M.

Waldorf-Astoria Jade Room

ELECTRONIC COMPONENT PARTS

Chairman: GHBERT B. DEVEY, Sprague Electric Co., North Adams, Mass.

27.1. An Evolution Is Coming

RICHARD DEWITT, DSDD R & E, The Pentagon, Washington, D. C.

New concepts of performing electronic functions have been exploding throughout the electronics industry. Many of these have merit; some will fall by the wayside. All have one thing in common—they represent dynamic and progressive thinking. We as engineers will have to become conversant with these new ideas, watch their progress, and be flexible enough to utilize those that will enhance our technological position.

Many enthusiastic proponents claim that our present concepts and efforts are obsolete. Do not become alarmed, just prepare to adapt or develop these concepts to your work as the technology advances. In other words, be alert to take advantage of the advancing state of the art.

These new concepts are known by many names: microminiaturization integrated components, complex components and many others. These names will be examined and analyzed. and specially developed resistors, capacitors, and other elements. Recognizing that as circuits are made more compact (at any given power handling capability) the natural result of increasing power density is increasing temperature, Thermionic Integrated Micro-Modules (TIMMS) have been designed to operate at 600°C. This temperature is achieved by surrounding the microminiature electronic components with suitable high-temperature insulation to contain the power dissipated in the tubes and resistors. Thus, the electrical power supplied to this type of circuitry is used very efficiently. The new electronic tubes and circuit elements are made from ceramic and metal parts particularly suited for high-temperature operation and are brazed together into stacks which yield a component density in the order of

10⁶ components per cubic foot. This paper reviews the design concepts mentioned above and presents an up-to-date report on the component development work which is being carried on at General Electric. Performance characteristics of these components and some of the T1MMS application considerations will also be covered.

This work is still in the developmental phase. Several applications of TIMMS circuitry have been considered in detail and have posed interesting electrical and thermal design problems. This paper will discuss the design approach to the solution of one of these problems to illustrate a method of solution found to be usable.

Thermionic Integrated Micro-Modules are not only small physically, but can be operated at low input voltages completely compatible with current transistorized circuitry. The speed of response of these vacuum tube circuits is very high. Many of the electrical connections required by the electronic function are made within the stack arrangements. Additionally, all the components, both passive and active, are evacuated and continuously gettered and are radiation tolerant, which makes them particularly suited for missile and other military device applications. The necessary thermal enclosure can be constructed in a number of different ways to optimize for minimum over-all system weight or volume as the circumstances require.

This basically new form of microminiature electronics is becoming an important new approach to the solution of the many limitations imposed by the power handling and temperature restrictions of solid-state devices. This paper should serve as an introduction to the TIMMS concept of electronics.

27.5. Microcircuitry—A Practical Technology for Reliable Microminiaturization

F. P. GRANGER, JR. AND J. G. SMITH, Varo Mfg. Co., Inc., Garland, Tex.

Microcircuitry is the technology for fabricating complex electronic circuits directly from fundamental particles, eliminating components and connections. Microcircuitry research into fundamental properties of matter and the means of combining molecular particles to obtain any predetermined transfer function is progressing toward ultimate reliability, minimum size, and minimum cost of electronic devices.

The present state of the art of microcircuitry permits the design and fabrication of a wide variety of reliable devices with effective parts densities of from one to ten million parts per cubic foot.

blocks and their present status in the molecular electronics program will be discussed. Functional electronic blocks (FEBS) are units of material grown, developed, and/or processed to perform basic electronic operations such as amplification, detection, and switching. How FEBS are being developed to satisfy the various functional requirements of advanced Air Force electronic equipment will be examined, examples will be given, and actual photographs will be shown.

The impact of this new technology on electronic component parts manufacturers will be considered. The future role and position of these manufacturers in the molecular electronics program and their possible contributions towards solving some of the current problems will be brought out.

Some thinking will be generated regarding the future employment of FEBS as inventory items for use in equipment developments vs the independent development of blocks for specific equipments utilizing molecular electronic technology, Example: The system engineer in conjunction with his molecular electronics team may want to design his own solid block of molecular electronics rather than put together the FEBS available from the inventory. The available FEBS may not completely satisfy the system requirements for frequency, stability, etc.

27.3. Electronic Progress —Circa 1960

Col. L. J. D. ROUGE AND G. M. R. WINKLER, U. S. Army Signal R&D Lab., Ft. Monmouth, N. J.

The frequent reference to the "galloping technologies of our day" is aimed at the flood of new concepts, devices, techniques, and materials pouring out of organized electronic research and development. In this connection, the Army's newer contributions in the fields of quartz crystal research, thin film capacitors, microwave ferrites, tiny reed relays, solid circuit devices, magnesium batteries, new cathode techniques, kilomegacycle "micro-mesa" and multiwatt "macro-mesa" transistors, etc., are significant parts of this vigorous stream of progress in electronics. This survey paper will discuss and illustrate these and other recent highlights of the Army components and materials R&D programs.

27.4. The Thermionic Integrated Micro-Module Program

C. G. CHILDS, A. P. HAASE, M. W. HAMILTON, AND R. M. HUGHES, General Electric Co., Owensboro, Ky.

In recent years component developments have made possible rapid progress in electronic circuit microminiaturization. Engineers of the General Electric Company have developed a new approach to compact efficient electronic circuitry which employs tiny thermionic tubes

^{*} Sponsored by the Professional Group on Component Parts. To be published in Part 6 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

25.2. Optimum Coincidence Procedures for Detecting Weak Signals in Noise

JACK CAPON, Federal Scientific Corp., New York, N. Y.

Previous investigators have found the optimum coincidence detector for particular detection problems by a point-by-point graphical procedure. Here it is shown that if the signal is weak compared to the noise, an analytic method can be used to determine the optimum coincidence detector. This method is based on Pitman's concept of asymptotic relative efficiency. The optimum coincidence procedure is compared to the optimum Neyman-Pearson observer. In the case of envelope detection of a sine wave in narrow-band normal noise, it is shown that the optimum coincidence procedure requires a signal-to-noise power ratio only 0.955 db greater than the optimum Neyman-Pearson method requires.

25.3. A General Theory of Signalto-Noise Improvement, with Application to the Visual Detection of Weak Signals

NORMAN S. POTTER, The W. L. Maxson Corp., New York, N. Y.

A new theory of signal-to-noise ratio improvement through readily implemented techniques is developed. It is applied to an empirically derived statistical model of human observer response to weak signals in noisy environments. An optimal postdetection operator, which is a function of the probability density functions of signal-plus-noise and noise alone, operates on the detector output prior to the display generator to maximize the contrast value of the signal to the mean surround. The technique is shown to yield large improvements in detection probability. The theory is applied to several highly recurrent models of signal and noise.

25.4. Information Rates in Photon Channels and Photon Amplifiers

THOMAS E. STERN, Dept. of Elec. Engrg., Columbia University, New York, N. Y.

A discrete "photon channel" is studied. The maximum entropy source under an average power constraint is determined. The transmission rate through a channel with Poisson signal and additive Poisson noise is calculated and compared to that for an additive Gaussian channel. The two are shown to be similar in the "continuous" region and to differ in the "quantum" region.

These results are applied to amplifiers. It is shown that the communication efficiency of a "physically ideal" amplifier of the maser type operating at 0° K approaches 50 per cent asymptotically for large average numbers of photons and zero asymptotically for small numbers of photons.

25.5. An Aspect of Information Theory in Optics

HIDEYA GAMO, IBM Res. Center, Yorktown Heights, N. Y.

One of the most characteristic features of optical systems is that the nature of optical images is dependent upon various conditions of illumination, namely, coherent, partially coherent, and incoherent illumination. By using the response function, sampling theorem, and intensity matrix for optical images, the limit of resolution among neighboring objects is shown to be essentially the same under various conditions of illumination. The amplitude and phase information of an object under various conditions of illumination, derived by using the elements of intensity matrix for images, which can experimentally be determined, is also discussed.

SESSION 26*

Wed. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Astor Galley

BROADCAST AND TELEVISION RECEIVERS

Chairman: W. L. DUNN, Admiral Radio Corp., Chicago, Ill.

26.1. Reduction of Modulation Defocusing in Television Picture Tubes

JOSEPH HOEHN, Allen B. DuMont Labs., Clifton, N. J.

A principal factor causing deterioration of picture quality in a television display is the effect of spot blooming on the highlights. This effect is caused by the increase in space charge repulsion in the electron beam at high currents. An increase in space charge results in an enlarged spot diameter under optimum focused conditions and, of even greater significance, the electron lens must be made stronger to achieve optimum focusing. Thus, under fixed focus conditions, as in all commercial TV sets, the electron spot is both enlarged and defocused at high beam currents.

A reduction in this modulation defocusing has been achieved through the design of a special prefocusing system that is formed in the region of the screen grid. The prefocusing field acts to control the rate of growth of beam diameter with increasing current in a proportion necessary to compensate for the increased force of repulsion (space charge).

26.2. Recent Developments in Scan Magnification

N. PARKER, I. CSORBA, AND N. FRI-HART, Motorola Inc., Chicago, Ill.

A brief review of the concept of scan magnification and the known ways by which it can be accomplished in a cathode-ray picture take is given. The idea of the negative lens in electron optics is discussed and the use of this type of lens in electron beam refraction is described. The paper will include a development of the relationships in negative electrostatic lenses which describe the magnitude of scan magnification and the effect on spot size. The considerations are given for such potential problems as secondary emission, lens distortions, and raster shadowing. The physical problems in the construction of a cathode-ray tube which provides scan magnification are described. The paper will conclude with a demonstration of an operating television picture tube incorporating a scan magn fier.

26.3. Noise Figure Performance of VHF Transistors and Tubes at Various Operating Conditions

J. F. BELL AND L. E. MATTHEWS, Zenith Radio Corp., Chienne W

Chicago, Ill.

A noise figure measurement technique which quickly yields first-stage noise figure and gain is described. The results of such measurements on a variety of VHF transistors are presented as contours or constant noise figure and constant gain on the collector current-voltage plane. A comparison of similar measurements on vacuum tubes is given.

26.4. A New High-Performance AM/FM Transistorized Portable Receiver

B. J. MILLER AND E. A. SNELLING, Zenith Radio Corp., Chicago, Ill.

A description of a transistorized AM/FM receiver embodying new circuit features which include AFC, combined AM/FM 1F strip, efficient AGC, and an RF stage optimized for both AM and FM.

26.5. Filter-Phaser AM Stereophonic Receiver

A. A. GOLDBERG AND A. KAISER, CBS Labs., Stamford, Conn.

This describes a receiver for a system of stereophonic broadcasting, which employs conventional amplitude modulation for L+R and quadrature suppressed carrier sidebands for L-R.

An envelope detector is used to demodulate the L+R signal and a synchronous detector, the L-R signal. An IF half-lattice crystal filter extracts the carrier for reinsertion at the synchronous detector. The carrier phase shift through the crystal filter is used to control the frequency of the local oscillator by means of a

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The tube operates at an efficiency in excess of 50 per cent over a bandwidth of at least 3 per cent. The gain of the device is in the range of 10-13 db. In addition to being the first high-power CW application of the platinotron principle, the tube makes use of new techniques for handling anode and cathode dissipation which greatly extend the power handling capability of the continuous-cathode crossed-field device.

23.2. High-Power L-Band CW Traveling-Wave Tube Amplifiers

R. STRAUSS AND J. MCCAMMON, Sperry Electronic Tube Div., Gainesville, Fla.

During the past few years advances have been made in the field of high-power (300 watts) CW traveling-wave tube amplifiers. This paper presents the results of a comprehensive development and refinement program on two complementary tube types. The mechanical and electrical characteristics of both types are presented. The design concepts are described. Included in the discussion will be some novel characteristics pertaining to harmonic power generation and the effects of load on tube operation.

23.3. The Effects of Magnetic Focusing Fields and Transverse Beam Velocities on Spurious Oscillations in Backward-Wave Oscillators

LOREN L. MANINGER, Sylvania Products Inc., Mountain View, Calif.

One of the major problems encountered in backward-wave oscillators is spurious oscillation. These oscillations appear to be of the form $nf + m(f + \Delta f)$, occurring on both sides of the main carrier. Theoretical considerations have shown that spurious oscillation is effected by magnetic fields and transverse electron motion. These effects have been verified experimentally on tubes in which transverse magnetic fields and transverse beam velocities have been introduced. The results of these experiments on spurious oscillation level, the ratio of the starting currents of spurious oscillation to the main mode starting currents, beam transmission and power output are presented. The proper selection of design parameters should result in minimized spurious conditions.

23.4. The Design and Performance of a Commercial Ammonia Maser Oscillator

S. HOPFER, Polytechnic Res. and Det. Co., Inc., Brooklyn, N. Y.

Some of the numerous engineering problems associated with the design of a completely selfcontained 12X19 inch rack-mounted ammonia maser oscillator will be discussed. In particular, the practical solutions to the problems of cavity pulling, line shape, Doppler shift, maser tuning, and maser sealing-off will be discussed. The RF circuitry associated with the utilization of the ultrastable maser signal, as well as other auxiliary equipment included in the design for safe maser operation will be described. The paper will conclude with a presentation of performance data obtained from three maser models.

23.5. Extended-Dynamic-Range Traveling-Wave Tubes

J. KLIGER AND E. J. DOWNEY, Hughes Aircraft Co., Culver City, Calif.

The "successive-signal-removal" principle has been used successfully in the past to increase the dynamic range of intermediate-frequency amplifiers and of traveling-wave tube amplifiers having video outputs. A brief review of the principle and its application is followed by a résumé of recent work done to increase the range of traveling-wave tubes operating entirely at the microwave level. Increases in input-saturation levels of about 20 db have been realized in experimental S- and X-band amplifiers. These increases have resulted in dynamic ranges between 60 and 70 db in 15-db noise figure tubes. The small-signal gains of the devices were about 20 db. Design principles and the potentialities and inherent limitations of the method are discussed.

24.3. Reconnaissance—Radio, Radiation, Infrared, Optical

COL. B. S. PULLING, Deputy for Radiation Warfare Support, WPAFB, Ohio.

24.4. Design for Survival (Personnel and Material)

H. STRUGHOLD, School of Aviation Medicine, Brooks AFB, San Antonio, Tex., AND T. C. HELVEV, Radiation, Inc., Melbourne, Fla.

24.5. Communication Relaying

W. H. RADFORD, Mass. Inst. Tech., Cambridge, Mass., AND J. R. PIERCE, Bell Telephone Labs., Murray Hill, N. J.

The panel includes many of the nation's leading experts in space technology. The members will review current and future demands which will be made upon electronics as our knowledge and the use of space advances.

SESSION 24*

Tues.

8:00-10:30 P.M.

Waldorf-Astoria Grand Ballroom

PANEL: ELECTRONICS—OUT OF THIS WORLD

Chairman: ERNST WEBER, President, Polytechnic Institute of Brooklyn, Brooklyn, N. Y.

24.1. Intergalactic Data

L. V. BERKNER, Associated Universities, New York, N. Y.

24.2. Weather Forecasting and Control

Admiral L. deFlorez, USNR, Englewood Cliffs, N. J., and M. TEPPER, Meteorological Satellite Programs, NASA, Washington, D. C.

SESSION 25*

Wed.

10:00 A.M.-12:30 P.M.

Waldorf-Astoria Starlight Roof

DETECTION THEORY AND APPLICATIONS TO PHYSICS

Chairman: PETER ELIAS, Mass. Inst. Tech., Cambridge, Mass.

25.1. Estimation of Doppler Shifts in Noise Spectra

PETER SWERLING, The R.1 ND Corp., Santa Monica, Calif.

A sample function of a stationary Gaussian process $\{N(t)\}$, having zero mean, is observed for a finite time. The power spectrum of $\{N(t)\}$ is assumed to belong to a family of functions $\{F_h(\omega)\}$, where $F_h(\omega) = F[\omega(1+h)]$, and $F(\omega)$ is a specified even, non-negative, square integrable function of ω . The problem is to estimate *h*.

Formulas are derived for the mean and variance of estimates belonging to a certain class of estimates of *h*. Modifications of these formulas are given for the case where $F(\omega)$ is imperfectly known.

An illustrative numerical example is given. A possible application to space-flight navigation is pointed out.

March

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sonic depth sounds in the body, similar to work done on reflection of pathological bone. The attempt is indicated for development of a clinical tough inexpensive method of predicting fetal and maternal mortality, so that necessary steps may be taken to reduce this danger. Recourse is made to the use of small-size permanent-type magnetic speakers with the paper cone attached to a solenoid coil, floating in the narrow air gap of a permanent magnet. A 45-ohm best impedance match was used, and the interpretation of sounds from an artificial chamber were measured, as were actual clinical experiments conducted in a large city hospital, cases then undergoing Caesarian section, verifying the suspicions pointed to by the electronic transducer. Medical possibilities are indicated, and the future use of such an instrument, now being questioned as Laennec with his first earscope, may well reach the stage of the common stethoscope.

21.4. Use of a High Sensitivity Capacitance Pickup in Heart Sound Research

D. GROOM, M.D. AND Y. T. SHIVO-NEN, Medical College of South Carolina, Charleston, S. C.

The authors have been impressed with a need for electronic techniques capable of detecting and recording heart murmurs of very low intensity-sounds at and below the threshold of stethoscopic audibility. To accomplish this necessitates a pickup having high sensitivity and wide range, and extremely low levels of ambient noise both in the recording system and in the environment. They have attempted to devise such a pickup. It is essentially a capacitance transducer which does not rely on air conduction of sound and which can utilize the body surface itself as one electrode of the capacitor. Clinical and experimental tests of this capacitance pickup indicate that it is uniquely suited to the recording of cardiovascular sounds.

SESSION 22*

Tues.

2:30-5:00 P.M.

Coliseum Marconi Hall

DESIGN OF EQUIPMENT RELIABILITY

Chairman: E. J. BREIDING, IBM Corp., Kingston, N. Y.

22.1. Safety Margins Established by Combined Environmental Tests Increase Atlas Missile Component Reliability

C. C. CAMPBELL, Convair Astronautics, San Diego, Calif.

The Convair Astronautics Reliability Test Program is a systematic search for weaknesses in missileborne equipment. All tests are conducted in laboratories which have capability for simulating equipment operating conditions and subjecting the equipment to environmental severities at and beyond those expected in flight. Testing beyond design requirements is done to establish a margin of safety and to create a basis for determining the comparative reliability of all missileborne components. After completing the first three phases of this program, 203 component types, consisting of 980 individual specimens, have been evaluated. This paper will discuss the techniques used in implementing the program, the results achieved to date, and the methods used for effecting corrective action on detected component weaknesses.

22.2. Segregating Subsystem Errors of a Transistor Magnetic Circuit

WALTER R. KUZMIN, Packard Bell Electronics Corp., Los Angeles, Calif.

This paper describes how a statistical design of experiment was planned to segregate the errors contributed by each subsystem of an electronic circuit composed of the power supply, magnetic cores, and transistorized amplifier. The experiment is described in detail and mentions how small random samples can be utilized efficiently to determine the cause of extreme variability. Upon completion of the experiment, charts are presented which compare the contribution to error by each of the aforementioned individual subsystems. The results also show some startling effects of temperature on magnetic toroids, transistorized amplifier, and power supply.

22.3. The Statistical Analysis of Redundant Systems

FRED MOSKOWITZ, Rome Air Dev. Center, Rome, N. Y.

The statistical basis for redundant systems which can be characterized by a probabilistic graph or redundancy network is explored. It is shown that in addition to the more commonly used survival probability function or reliability and its inverse, the cumulative failure probability functions, such related statistical func-tions as the "hazard," the "safety," and the "mortality distribution" are useful in the statistical analysis of systems. The interrelationship between these functions are discussed and simple formulas are given by means of which it is possible to derive all the remaining functions when one function is known. The use of the exponential, normal, logarithmic normal, and gamma distributions is illustrated and their range of applicability is discussed. Methods are given for deriving or estimating the confidence intervals for given confidence levels of a redundant system when the redundancy network and the confidence interval of the system elements or components are given. In general, it is shown how it is possible to derive the statistical behavior of a redundant system from the statistical behavior of its component parts through the redundancy network and its associated redundancy function.

22.4. Some Results of an Early Reliability Program

RALPH E. KUEHN, IBM Corp., Owego, N. Y.

A reliability life test was conducted at several phases of the life cycle of a bombing navigation system to determine compliance with a design requirement of 5000 hours mean time to failure on electron tube modules. Statistically significant results are discussed as well as the cost and economic savings realized by the customer.

The use of a detailed analysis of each component part application and the effect of part modes of failure on system performance are demonstrated by a discussion of relay malfunctions occurring in field operations. Practical and valuable analysis of field failure data is described.

22.5. Maintainability Profile Analysis

H. E. THOMAS, J. SOUKUP, AND W. BROBST, *ITT Labs.*, *Nutley*, N. J.

This material presents an evaluating system for specifically quantitizing the maintainability of operating electronic equipment. It seeks to definitize and place measurable figures of merit on equipment designs by establishing a series of key and universal maintainability factors.

When these factors in an equipment undergoing design review are quantitatively applied, they may be represented in a two-dimensional manner upon a histogram type of graph. From a summation of the resultant graph's profile, an average over-all maintainability index may be calculated for comparison and evaluation purposes.

SESSION 23*

Tues.

2:30-5:00 P.M.

Coliseum Morse Hall

MICROWAVE TUBES

Chairman: MARVIN CHODOROW, Stanford University, Stanford, Calif.

23.1. High-Power CW X-Band Amplitron

WILLIAM C. BROWN AND GERALD PERLOFF, Raytheon Co., Waltham,

Mass.

The platinotron principle has been applied to an *X*-band CW amplitron which nominally operates at 2.5 kw of power output but which has given as much as 8 kw of power output.



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SESSION 20*

Tues.

2:30-5:00 P.M.

Waldorf-Astoria Sert Room

AUDIO AND BROADCAST AND TELEVISION RECEIVERS

Chairman: DANIEL W. MARTIN, The Baldwin Piano Co., Cincinnati, Ohio.

20.1. The Present Status of Stereo Broadcasting

C. G. LLOYD, General Electric Co., Auburn, N. Y.

20.2. Receiver Design Considerations for Stereophonic FM Multiplex Broadcasting

C. G. EILERS, Zenith Radio Corp., Chicago, Ill.

Circuits within the receiver are analyzed relative to stereophonic crosstalk. Various circuit approaches for demodulating the subcarrier are discussed and compared on the basis of complexity, stability, and performance. Effects of multipath and ignition pulse interference on stereophonic receiver performance are described.

20.3. The Percival Stereophonic System

W. S. PERCIVAL, Res. Labs., Electric & Musical Industries, Ltd., Hayes, Middlescx, England.

The purpose of the system is to provide a stereophonic signal for radio transmission within a bandwidth only slightly greater than that required for ordinary monophonic transmission. The left and right microphone signals are combined to form a single AF signal which is available for ordinary monophonic reception. They are also processed in such a way as to emphasize the beginnings of sounds and then utilized to form a signal, with a bandwidth of 100 cps carrying the directional information only. This directional signal is caused to modulate a subcarrier at a power level much lower than that of the AF signal. In the receiver the AF signal and the directional signal are applied to a Hall multiplier unit giving the required left-loudspeaker signal as one output and the right-loudsneaker signal as the other.

20.4. A Continuously Variable Wireless Remote Control for Stereophonic Phonographs

A. A. GOLDBERG AND A. KAISER, CBS Labs., Stamford, Conn.

This describes an ultrasonic remote control for a stereophonic phonograph that provides continuous control of volume and stereophonic balance as well as record reject. The design incorporates techniques for immunizing the system against spurious signals.

Control of the separate functions is accomplished by transmitting and receiving one of three CW ultrasonic signals. Noise immunization is inherent in the logical circuitry that precludes operation if more than one frequency is received.

The transmitter consists of a transistor generator and a ceramic transducer. The receiver employs a similar transducer and a broadbanded 3-stage amplifier-limiter driving frequency selective circuits to trigger one of 3 thyratrons. Two of the thyratrons form part of a bridge to control the rotation of dc motors coupled to the volume and balance potentiometers. The third thyratron controls the record reject solenoid.

20.5. Automatic Stereophonic Phaser

B. B. BAUER, A. A. GOLDBERG, AND G. POLLACK, CBS Labs., Stamford, Conn.

With the advent of sterophonic records and broadcasting, there is a need for a device that can sense the phasing of the Left and Right signals and automatically make corrections if necessary.

The CBS Laboratories' Automatic Stereo Phaser bridges the Left and Right program lines, linearly amplifies each signal and then converts them to L+R and L-R signals by means of a transformer matrix. The L+R and L-R signals are separately rectified and the resulting dc is applied to a mechanical flip-flop. Correct phasing results in a greater L+R signal as compared to the L-R signal. If the reverse appears, the program lines are automatically rephased.

SESSION 21*

Tues.

Coliseum Faraday Hall

2:30-5:00 P.M.

THE HUMAN AS ORIGINATOR OF SIGNALS AND SCHEMES

Chairman: H. H. ZINSSER, M.D., New York, N. Y.

21.1. Implantable Cardiac Pacemakers

WILSON GREATBATCH, Electronics Consultant, Clarence, N. Y.

The state of the art of transistor electronics is approaching the point where serious consideration can be given to implanting electronic devices in the human body for periods of unattended service of up to five years. The cardiac pacemaker represents an ideal application for this class of equipment. This paper will discuss the requirements of such a pacemaker. A sample unit will be demonstrated that has actually been implanted in an experimental animal and has been in successful operation for over three months.

21.2. Detection and Analysis of High-Frequency Signals from Muscular Tissues with Ultra-Low-Noise Amplifiers

W. K. VOLKERS, M.D., Millivac Instruments Div., Cohu Electronics, Schenectady, N. Y., AND W. CANDIB, M.D., St. Claire Hospital, Schenectady, N. Y.

The recent perfection of low-noise amplifiers, such as the Hushed Transistor Amplifier, makes it possible to investigate exceptionally weak electromagnetic signals which are generated by the human body. Before such lownoise amplifiers were developed, the only effective means of suppressing amplifier noise was a rather drastic reduction of bandwidth, and investigations quite naturally confined themselves mostly to the low-frequency region (cardiographs and encephalographs). Now available, greatly improved, wide-band lownoise amplifiers enable us to study low-level signals generated by the body, over much wider frequency ranges than before. For instance, the deltoid muscle (in the shoulder), or the gastrocnemius muscle (calf muscle of the leg) generate sharply spiked signal waves when contracted; these waves contain frequency components up to well over 10 kc.

The paper describes the test equipment which combines a Hushed Transistor Amplifier with a new, especially developed, narrow IF band heterodyne postamplifier which is used to scan the frequency spectrum of muscle voltages. Typical spectra for various muscles are discussed, also pathological deviations in muscular structure (such as scleroderma, or thickening, scleroedema, or retention of water in tissue, and ischemic conditions, or diminished blood supply). These abnormalities influence the frequency spectrum of the muscle. Therefore, the low-level, high-frequency measurements of muscle voltages described in this paper promise to become an important tool for diagnosis of muscular diseases in the future.

21.3. The Stereo-Dynamic Aspects of Fetal Auscultation and Its Application to Medical Diagnosis

L. E. GARNER, JR., Silver Spring, Md., AND F. D. NAPOLITANI, M.D., Mount Vernon, N. Y.

This paper reveals the use of mono- and diphonic transducer pickups to the uterus and its contents. It also reveals the effects of picking up

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SESSION 18*

Tues.

2:30-5:00 P.M.

Waldorf-Astoria Astor Gallery

INDUSTRIAL ELECTRONIC INSTRUMENTATION

Chairman: JOHN J. GRAHAM, RCA, Camden, N. J.

18.1. An Inquiry into the Computer Automation of Supermarkets

RONALD R. SEGEL, Thompson-Ramo-Wooldridge, Inc., Los Angeles, Calif.

This is the age of automation. There is one facet of this that touches each one of us and that is the automation of supermarkets. Anyone who has stood for a long time at a checkstand waiting to be checked out will attest to the need. This paper has to do with machine recognition of the price of a supermarket item or of a code number upon the merchandise. The machine will take the place of the person who now operates the cash register and in addition will aid in inventory control. This paper covers the technical and philosophical aspects of an automated system. It points out several proposals that have been tried and some that are new.

18.2. Automatic Testing and Calibration of Central Air Data Computer

H. LANGENTHAL, Bendix Aviation Corp., Teterboro, N. J.

An automatic test set for the production testing and calibration of Central Air Data Computer System is described. A punched tape programming device is used to select the inputs, which consists of an electrical signal equivalent to an angular position of an autosyn to a resolution of 0.1°. A second punched tape programming device is used to select and channel on a parallel entry printer a maximum of 10 switch closures, 10 preselected capacitor values, 30 potentiometers, 20 autosyns, and 2 digitizers. The recorded information is displayed on a preprinted data sheet showing upper and lower tolerances. Calibration time will be cut approximately in half.

18.3. Electronics in Agriculture

F. C. JACOB, University of California, Davis, Calif.

Factors are outlined which favor and oppose the use of electronics in agriculture. Agriculture is compared with other industry regarding the need for automation. Current applications of electronics directly to agriculture are reviewed. Comments are made on experimental work now in progress on electronic equipment tions are offered.

Abstracts of Technical Papers

18.4. The Shawmeter—An Electronic Two-Color Pyrometer

VINCENT G. SHAW, Shaw Instrument Corp., Latrobe, Pa.

This paper briefly discusses the advantages of two-color pyrometry over some better known, temperature measuring devices. The electronic aspects of the instrument are covered in considerable detail. Potential sources of trouble in an instrument of this type and how they have been avoided are discussed. A few typical applications involving the Pyrometer in some industrial automation problems are reviewed.

The author attempts to show how the correct combination of electronic and physical principles can result in a device which is ideally suited for the ever-growing trend towards industrial automation.

SESSION 19*

Tues.

2:30-5:00 P.M.

Waldorf-Astoria Jade Room

BROADCASTING-II

Chairman: WILLIAM HUGHES, Iowa State College, Ames, Iowa.

19.1. Some Engineering Aspects of Video Tape Recording Production

EDWARD E. BENHAM, KTTV, Inc., Hollywood, Calif.

Hollywood is the focal point of film and live TV productions. It should become a major video tape production area. There is much activity along these lines, and this paper will describe some of these activities and the major role that engineers will play.

The approach of VTR productions should be patterned neither for live television or film, but rather somewhere between the two. In general, video tape recording productions can be divided into four sections: preproduction planning, production, postproduction editing, and laboratory work. In each, careful engineering planning can save time, facilities, and money.

The use of VTR as an economical production tool in transferring to film for release has increased the over-all usage of video tape, but requires that the end medium of release must be considered in the original productions.

The laboratory portion of the production is where the prints are made, rerecording either for video tape recording or film transfers, and final quality improvements are worked into the final release.

Many engineering details of the various steps will be described and the necessity of careful consideration of the total problem at the outset of any production will be discussed.

19.2. A Modern TV Transmitter Plant Input System

JOSEPH L. STERN, CBS-TV, New York, N. Y.

As the final link in the television transmission system the transmitter plant has an extrenely high responsibility. To meet this responsibility it must have an input system that can act as a nerve center and be capable of simple signal processing, continuous monitoring, automatic alarm signaling, and instant emergency action. As part of a program of plant updating and quality improvement, a completely new and simple transmitter plant input system has been designed. The key features of the system are:

1) "Human engineered" controls,

- 2) Remote controlled emergency by-pass circuits,
- 3) Simplified jack-fields,
- 4) Push button audio and video monitoring,
- 5) Automatic homing for normal monitor circuits.

System design, circuit and operating details are described.

19.3. A Special Effects Amplifier for Noncomposite or Composite Monochrome or Color TV Signals

RALPH C. KENNEDY, NBC, New York, N. Y.

A switching circuit has been designed to produce a doublet impulse transition of 0.05 μ see. The problem of clamping a color signal diring the burst interval by means of crystal giodes is discussed and a solution presented. Nonlinear amplification of the switching data prior to regenerative clipping has been found to permit dependable switching with much smaller brightness changes.

19.4. Remote Control of TV Microwave Equipment

JOHN B. BULLOCK, RC.4, Camden, N. J.

Recent actions by the FCC have made it possible for privately owned and operated microwave links to be constructed by television stations. The need for remote control of the systems has brought about the development of some novel techniques for input switching, system reversal, and the transmission of control information. Hardware designed for these functions will be described.

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15.4. A Mathematical Analysis of the Performance of the ATC Radar Beacon System

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A. ASHLEY AND F. H. BATTLE, JR., Airborne Instruments Lab., Mineola, N. Y.

A mathematical model representing the Air Traffic Control Radar Beacon System operation in the current New York environment has been developed to evaluate the effects of 1) changes in system parameters, and 2) incorporation of various combinations of system improvements. The combined effect of all stations on the reply rates from individual aircraft in the area is calculated, as well as the total reply information available for display and utilization at the Idlewild en route beacon station. To indicate the optimum future configurations, the computations are performed for combinations of 1) the basic system and three different sidelobe suppression techniques, 2) receiver subtraction, 3) single and double defruiting, and 4) several decoder settings. Various settings of dead-time, suppress-time, and reply-limit are imposed on each of the combined systems. Conclusions and recommendations are presented.

SESSION 16*

Tues.

10:00 A.M.-12:30 P.M.

Coliseum Morse Hall

BROADENING DEVICE HORIZONS

Chairman: M. E. HINES, Bell Telephone Labs., Murray Hill, N. J.

16.1. Masers

J. W. MEYER, M.I.T. Lincoln Lab., Lexington, Mass.

16.2. Variable Reactance Devices

B. SALZBERG, Airborne Instruments Lab., Mineola, N. Y.

16.3. Tunnel Diodes

H. S. SOMMERS, JR., RCA Res. Lab., Princeton, N. J.

16.4. Functional Devices

W. A. ADCOCK, Texas Instruments Inc., Dallas, Tex.

The advent of new active elements on the device horizon and the combination of several elements in more complex functional devices promise to broaden the vista of the electronics industry. The panel of speakers will present coordinated talks aimed at a broad coverage of these new devices. This will include a description of the fundamental aspects of the devices, a summary of the present state of development and an indication of some of the device development and application possibilities of the future. The chairman will present a brief summary and the final part of the meeting will consist of a panel discussion and questions from the floor.

SESSION 17*

Tues.

2:30-5:00 P.M.

Waldorf-Astoria Starlight Roof

RADAR AND CODING THEORY

Chairman: GEORGE L. TURIN, Hughes Aircraft Co., Culver City, Calif.

17.1. Sequential Procedures in Radar Pretracking

MISCHA SCHWARTZ, Polytechnic Inst. of Brooklyn, Brooklyn, N. Y.

Two schemes for the preliminary tracking of a target in a combined and automatic detection-tracking radar are discussed and analyzed.

These serve as detection verification procedures, decreasing the number of times noise is erroncously tracked as a true target. They are sequential procedures in which a definite decision to track a target is withheld until the signal measured exceeds the upper of two threshold levels.

The average number of radar scans beyond the detection scan is found to be between one and two, for a given high probability of discarding noise and correctly tracking a signal.

17.2. Detection Range Predicting for Pulse Doppler Radar

S. A. MELTZER AND S. THALER, Hughes Aircraft Co., Culver City, Calif.

A mathematical model of sufficient flexibility to describe most pulse Doppler radar search systems is constructed and used to predict detection ranges. It is applicable to situations where thermal noise or sidelobe clutter limits detection range. The sidelobe clutter is assumed to be statistically similar to thermal noise. The model allows the variation of most of the important radar parameters. The receiver is assumed to have a number of rectangular, nonoverlapping range gates followed by narrowband, Gaussian-shaped Doppler filters and then a square law envelope detector followed by a post-detection filter consisting of one or two stages having exponential weighting functions. Single scan and cumulative probabilities of detection are calculated for both steady and scintillating targets.

17.3. The Search Efficiency of the Sequential Probability Ratio Search Radar

GLENN W. PRESTON, General Atronics Corp., Bala-Cynwyd, Pa.

This report describes a method of radar signal detection and search in which signal detection is accomplished by a sequential probability ratio detector and the beam scanning is controlled by the detector. A theoretical improvement typically of 5 to 8 db is gained over nonsequential search radars which are otherwise identical.

17.4. Group Codes for Correcting Prescribed Error Patterns

ROBERT T. CHIEN, IBM Res. Center, Yorktown Heights, N. Y.

This paper studies the problem of constructing group codes for correcting error patterns. An equivalence relation is defined and error patterns are divided into equivalence classes. If a group code can be constructed for any pattern, a transformation technique will give group codes for all error patterns in the same class. For any error pattern, a necessary and sufficient condition for the existence of a correcting group code is given. Procedures for constructing group codes are illustrated for patterns which fulfill the condition. In the case r=3, out of more than 10⁷ matrices, only 20 equivalence classes were found.

17.5. Some Results on Best Recurrent-Type Binary Error-Correcting Codes

WILLIAM L. KILMER, Montana State College, Bozeman, Mont.

This paper treats error-correcting codes consisting of binary message sequences augmented by periodic insertions of parity check digits. The codes are studied in terms of a parity check matrix, *M*, an error pattern vector, and a parity check sequence vector. Problems concerning the codes can be reduced to questions about the columnar sum properties of *M*. Answers are given to many important questions regarding the burst-correcting properties of the codes. These answers are based on the specification of a mathematical group of binary sequence patterns. The codes can be designed to have easily instrumentable and highly efficient error-correcting properties.

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^{*} Sponsored by the Professional Group on Information Theory. To be published in Part 4 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

DANIEL I. WEINBERG, Astra, Inc., Raleigh, N. C.

Many enzymes can be assayed by measuring the rate of change of optical density of a component of the reaction mixture. The component may be the enzyme, its substrate, or a compound which interacts with the reaction products.

The device described in this paper was originally designed to assay adenosine triphosphatase in fluids. This enzyme splits its substrate, adenosine triphosphatase, releasing an acid which changes the optical density of phenol red, a pH indicator.

The device will store one hundred samples, add reagent to, and measure the rate of change of optical density of each, record this measurement and the sample number on a paper tape, and, when finished, turn itself off.

14.3. Biological Microwave Hazards

VICTOR T. TOMBERG, Biophysical Res. Lab., Elmhurst, N. Y.

The increase in microwave power for radars and industrial applications makes it necessary to revise former standpoints taken in physicotherapeutical applications of short waves and microwaves.

It is known that there are three classes of microwave field actions which have biological significance: 1) ordinary thermal effects, 2) specific thermal effects, and 3) electric effects. At low doses those actions are stimulative and beneficial, but at higher doses are harmful and destructive. Most of the hazards are produced by the specific thermal actions, even at low field intensities, of 1 watt per square cm and less. Specific thermal effects are often mistaken for electric effects, particularly in nonhomogeneous bodies or suspensions where parts or particles have a different dielectric behavior compared with the surrounding medium. Boundaries and interfacial layers act on the high-frequency field distribution, change the absorption pattern of the microwave field, and exert a focusing action which makes the value of the average field intensity meaningless so far as it concerns the safety requirements in some cases. Eyes and testicles are often vulnerable parts of the body. At low intensity fields we have to consider the voltage gradient and resonant phenomena more than the average field intensity.

Pearl chain effects are electrical effects mostly independent of the frequency. Because they only prevail when there is no significant thermal action, they do not constitute a biological hazard.

14.4. An Automatic Physiological Telemetry and Analog-to-Digital Conversion System

W. E. SULLIVAN AND C. A. STEIN-BERG, Airborne Instruments Lab., Deer Park, N. Y.; J. T. FARRAR, M.D., Veterans Administration Hospital, New York, N. Y.

Intraluminal pressure recordings provide valuable information concerning the motility within the gastrointestinal tract. Conventional methods for recording of pressure changes within gastrointestinal tract require the passage of tubes through the mouth, nose, anus, or through an artificial opening. An instrument has been designed to record the gastrointestinal motility with a minimum discomfort to the pament. This instrument consists of an ingestible pressure-sensitive telemetry capsule (pressure transensor) and an associated monitoring receiver.

The resulting recordings of gastrointestinal pressure are very complex and difficult to categorize. It was felt a general-purpose digital computer would greatly aid in the characterization of these complex waveforms. Thus, a system has been designed to take these gastrointestinal waveforms along with other physiological measurements, digitize this analog data, and record the analog data on magnetic tape in a format compatible with an IBM 704 computer.

14.5. Panel: Significant Variables in Biophysical Evaluation of the Human Under Stress

Chairman: C. D. RAY, University of Tennessee, Division of Surgery, Memphis, Tenn. Panel Members: L. CLARK, Medical College of Alabama, Birmingham, Ala.; MEMBERS OF THE STAFF OF COL. J. P. STAPP, Aero Medical Div., WPAFB, Ohio; O. H. SCHMITT, University of Minnesota, Minneapolis, Minn.

The panel will discuss the general physiological pattern to be followed in measuring the significant variables of the human under stress; the use of internal electrodes for tissue sampling; currently practical methods of gathering physiological data and what variables prove most fruitful under acute stress; and what physiological variables should be measured, including direct and integrated physiological functions.

SESSION 15*

Tues. 10:00 A.M.-12:30 P.M.

Coliseum Marconi Hall

MODERN APPROACHES FOR IMPROVED AIR TRAFFIC MANAGEMENT

Chairman: ALDEN C. PACKARD, FAA, Washington, D. C.

15.1. An Air Height Surveillance Radar (AHSR-1)

THOMPSON J. SIMPSON, F.1.1, Washington, D. C.

The Air Height Surveillance Radar (3-D) currently under development by the Federal Aviation Agency is a passive S-band receiving system which is used in conjunction with an airport surveillance radar as a source of target illumination. The system is capable of resolving two aircraft located fifty nautical miles distant from the antenna at the same azimuth and separated in elevation by 1000 feet. With a vertical aperture of 150 feet, the antenna produces about 115 separate receiving beams, the lower group having a beamwidth of about 0.19°. The output of each beam is fed to an individual receiver/beam-selector/height-computer_channel. A beam-selector matrix determines in which beam the target is located and with the aid of a computer determines the actual height of the target. The various height outputs then are used for special 3-D displays in addition to providing for three-dimensional automatic target tracking within the Data Processing Central System.

15.2. Automatic Ground-Air-Ground Communications for Control of Air Traffic

WAYMAN R. DEAL, FAA, Washington, D. C.

Voice radio channels forming the vital links for the control of civil and military air traffic are becoming seriously congested. Studies have shown that by 1970 air traffic control communications will quadruple the 1958 volume. Automation through the use of a data link offers a practical means to relieve congestion and provide growth potential to keep pace with the rapid acceleration in communication requirements.

The Federal Aviation Agency has developed the Automatic Ground-Air-Ground Communications System (AGACS) as an initial step in mechanizing routine communications. This experimental equipment is currently under evaluation at the National Aviation Facilities Experimental Center. The paper briefly describes the system and presents the results of the evaluation program to date.

15.3. Technical Research for Future Aviation Facilities

N. BRAVERMAN, W. W. FELTON, S. JUSTMAN, R. E. KESTER, MAJ. L. J. SCHAUB, USAF, AND A. WETTER, FAA, c/o NAFEC, Atlantic City, N. J.

This paper covers the role of technological research in the development of aviation facilities which can keep pace with the increasing number and expanding requirements of the users of the airspace. The methods for selecting new techniques for investigation are examined. The need for carrying on technological system synthesis simultaneously with the technical studies is explained. A discussion is included regarding the relationship of technical system synthesis to the study of the technical factors which are common to form a system of facilities. Reported are the objectives and progress to date of several research efforts which have been started within the Bureau of Research and Development of the FAA.

^{*} Sponsored by the Professional Group on Aeronautical and Navigational Electronics, To be published in Part 8 of the 1960 IRE INTLENATIONAL CON-VENTION RECORD.

tem compensating the audio system spectrum and equalizing disturbing resonance phenomena. This method has been successfully applied to the control of a vibration system spectrum where mechanical resonances at the exciter table produce high *Q* peaks and notches. The degree of spectral flatness reveals that adequate compensation is possible for typical resonance phenomena occurring in the frequency range above 500 cps. Expected improvement using narrower band-pass filter characteristics is presented.

12.5. An Analysis of Factors Affecting Recording Reliability and Digital Tape Recorders

KEN TAYLOR, Ampex Corp., Redwood City, Calif.

This paper develops the factors affecting recording reliability of digital magnetic tape equipment-primarily tape dropouts and dropout detection systems. The factors causing ing dropouts and the effects of dropouts are analyzed and classified. The relationship between packing density and recording reliability is presented, and methods are presented to determine maximum system reliability. The problems of head-to-tape contact are analyzed and means are presented to yield optimum head-to-tape contact in terms of reliability. The factor of maximum pulse resolution is also described, in terms of reliability. The conclusions drawn are that by sacrificing resolution for insensitivity of head-to-tape separation, the result should be a digital recording system which is operating at near optimum reliability for its specific packing density.

SESSION 13*

Tues.

Waldorf-Astoria Grand Ballroom

10:00 A.M.-12:00 NOON

ENGINEERING MANAGE-MENT-II

Chairman: JOHN R. RAGAZZINI, College of Engrg., New York University, New York, N. Y.

13.1. More Effective Engineering Proposals, One Key to Success

FRANK W. EVANS, JR., Dept. of the Navy, Washington, D. C.

Successful engineering proposals assure new contracts; new contracts sustain a healthy business; a healthy business generates profits with which to create new capacity and capability. The mission of an engineering proposal is to convince the prospective customer that your company's solution to his engineering problem is best, that your personnel and facilities are eminently qualified to accomplish his technical and production objectives.

The ABC's of successful proposals are Accuracy, Brevity, and Clarity which must be supported by technical honesty and design simplicity.

Engineering proposals must particularly demonstrate a high standard of technical excellence and accomplishment plus management interest in and support of the over-all program.

13.2. The Application of Closed Loop Control Techniques to Engineering Project Planning and Control

R. W. HAINE AND W. LOB, Bendix Aviation Corp., Teterboro, N. J.

In order to complete an engineering project within specification, allotted funds, and scheduled length of time, it is necessary to program the various phases of the project in great detail. Furthermore, a system for feeding back information for corrective action during the operation of the project is required to eliminate and avoid major departures from specifications, expenditures, and schedules.

This paper describes a system for planning engineering projects and controlling them by means of feeding back information on their progress periodically. The paper also shows the corrective measures used when deviations to the initial plans become apparent.

13.3. The Professional Register— A Program for Improving Engineering Management Visibility of Technical Capabilities

N. A. BEGOVICH, Hughes Aircraft Co., Fullerton, Calif.

The Professional Register recently installed in the Ground Systems Group of the Hughes Aircraft Company, Fullerton, Calif., is designed to serve three purposes:

- To provide technical personnel with an opportunity to record significant aspects concerning their background and career progress.
- To provide a mechanism for locating internal candidates for reassignment purposes.
- To provide information concerning the capabilities of engineering and scientific staff members for contract proposal and bidding purposes.

The program is implemented as follows. Each scientist and engineer will be provided an opportunity to specifically record background information, such as engineering specialty fields, education, and professional accomplishments. This information will be submitted on especially designed individual portfolios. Data are then translated via automatic data processing equipment and are made available for review on direct reading machine reports. As the need arises for personnel reassignment purposes or proposal capabilities, the master inventory is searched or summarized to provide the required information.

13.4. Management Control of Engineering Effort Through Graphic Methods

B. P. GOLLOMP, Bendix Aviation Corp., Teterboro, N. J.

This paper describes a graphic method for the preparation of engineering forecasts and budgets. Curves are employed to depict a project plan and its status. Other curves indicate the necessary planning corrections required as a project progresses. A third group of curves is used to indicate group or department status. Project and department budgets can be prepared from the aforementioned curves. The major advantages of the system are 1) individual and group professional abilities and talents, project complexity, and a variety of subjective factors are considered in the system, and 2) plans and status can be displayed and analyzed with a minimum of difficulty and training.

SESSION 14*

Tues.

10:00 A.M.-12:30 P.M.

Coliseum Faraday Hall

VARIED VIEWS OF MEDICAL ELECTRONICS

Chairman: II. H. ZINSSER, M.D. New York, N. Y.

14.1. Introductory Remarks— Training of Medical Engineers

H. II. ZINSSER, M.D., New York, N. Y.

The structure of medical engineering can be thought of as bridging links across multiple disciplines. Unlike biophysics, which depends primarily on mathematical or physical analytic techniques, medical engineering has its primary emphasis on synthesis, drawing on second-order derivations from at least two already derived disciplines, themselves one step removed from the basic sciences. As such a secondary derived area of learning and endeavor, it necessitates a wide range of basic knowledge and an open inquiry into techniques not as yet applied or in some cases even derived.

The essence is the applied synthetic art. As medicine and medical research become increasingly attractive as alternatives to purely analytical or destructive disciplines, it can be hoped that the synthesis of the practical and human oriented points of view that both professions share alike can merge mankind as painlessly and profitably as possible.

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^{*} Sponsored by the Professional Group on Medical Electronics. To be published in Part 9 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SESSION 11*

Tues. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria

Jade Room

BROADCASTING-I

Chairman: RAYMOND ROGERS, Westinghouse Broadcasting Co., Inc., KDKA, Pittsburgh, Pa.

11.1. Report on Geneva Radio Conference

W. H. WATKINS, FCC, Washington, D. C.

The Administrative Radio Conference of the International Telecommunication Union was held in Geneva, Switzerlaud, starting on August 17, 1959, for the purpose of revision of the Radio Regulations which were formulated at Atlantic City in 1947 and subsequently were annexed without change to the Buenos Aires Convention of 1952. An explanation of the organization of the conference will be given from both a national and an international point of view. The accomplishments of the conference will be discussed along with comments on how the new radio regulations will affect users of the radio spectrum in this country.

11.2. Future Possibilities for Film Room Mechanization

JAMES II. GREENWOOD, WCAE-AM/FM and WTAE-TV, Pittsburgh, Pa.

Complete mechanization of the film and slide operations in a television station is somewhat restricted by the characteristics of presently available equipment. This paper describes equipment which could be produced today, which would reduce the projectionist's work load and at the same time result in smoother programming as well as reduced chances for errors. The equipment components are equally suited to manual or machine switching.

On existing slide projectors, slides are shown alternately from two drums or disks, making it difficult to reuse a slide in a series or to add or delete slides. This paper discusses various methods of improved usage by the use of storage tube techniques. It also discusses methods and techniques for improved film projection operations and methods by which they might be accomplished.

11.3. Directional Antennas for Television Broadcasting

GEORGE II. BROWN, RCA, Princeton, N. J.

Desirable operating characteristics and other requirements of directional transmitting antennas, suitable for television broadcasting, are considered. Basic limitations on directivity are imposed by long-distance propagation phenomena. Previous experience with directional antennas used for television broadcasting is reviewed. A number of appropriate antenna types are suggested.

11.4. Service Area of an Airborne Television Network

MARTIN T. DECKER, Natl. Bureau of Standards, Boulder, Colo.

An educational television network using UHF transmissions from aircraft has been suggested as a means to provide improved classroom instruction to schools throughout the country. Initial airborne tests will begin in 1960. As a step in the evaluation of such a network, calculations have been made which describe the coverage from an airborne station in the presence of interference from other stations. The results should be useful in determining the equipment and channel requirements of a nation-wide network. These requirements will be compared with those necessary to provide equivalent coverage from ground-based UHF stations.

SESSION 12*

Tues. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Sert Room

AUDIO

Chairman: HARRY A. PEARSON, Sonotone Corp., Elmsford, N. Y.

12.1. A Plotter of Intermodulation Distortion

EDWARD F. FELDMAN, Panoramic Radio Products, Inc., Mount Vernon, N. Y.

For use mainly in audio systems, the instrument plots, on a CRT screen or chart recorder. the amplitude of the first-order difference frequency tone vs the lower excitation frequency in CCIF intermodulation distortion measurements. The unit furnishes two swept, equalamplitude tones with a selected audio difference frequency to excite the tested system. Readout is on a modified audio frequency spectrum analyzer which remains tuned to the difference frequency. The automatic excursion provides considerable test time economy in evaluation of such devices as loudspeakers which exhibit wide variation in intermodulation with excitation frequency. Conventional point-by-point techniques run the risk of missing critical conditions.

The test set is also capable of tracking thirdorder distortion, $(2f_1 - f_2)$. Residual 1M distortion is more than 80 db below the level of a single tone. A conventional swept spectrum analyzer and slave frequency response curve tracer are included in the complete system. Switching functions permit alternate displays on successive scans of any two of the modes of operation. For example, acoustic output and IM distortion vs frequency of loudspeakers may be plotted alternately for rapid evaluation.

12.2. Listener Ratings of Stereophonic Systems

HARWOOD B. MOORE, General Electric Co., Utica, N. Y.

Assorted stereophonic phonograph systems for the home were listened to under controlled conditions by numerous observers. The observers were given complete freedom to select alternately between system A and system B and asked to indicate a preference.

Each system was given the benefit of doubt by employing high quality amplifiers and speakers and each system was adjusted according to its theoretical description but with equivalent listening response characteristics.

This report describes the test conditions, test equipment, and test data obtained from the observers. Simple statistical logic has been applied to the data to present a phonograph system rating guide.

12.3. Calculation of the Gain-Frequency Characteristic of a Multimesh Transistor Amplifier Stage Using a Programmed Computer

D. E. BRINKERHOFF, General Motors Corp., Kokomo, Ind.

Electronic circuit designs are most commonly made by "analog" (or "breadboard") methods. This is particularly true in arriving at the desired gain and frequency response characteristics. An electronic circuit is essentially an analog computer and many parameters can be evaluated and quickly optimized by the "decade box method." On the other hand, a mathematical treatment of the equivalent multiple mesh circuits produces multiple simultaneous equations with multiple complex, frequency dependent coefficients which often require days instead of minutes to solve by hand calculation.

The introduction of the high-speed digital computer has made the mathematical approach not only feasible but in many cases faster than the breadboard method.

This paper will discuss the calculation of gain vs frequency of a transistor audio amplifier used in the output stage of an auto receiver. It will also investigate the effect of varying some of the parameters on the over-all gain and response of the stage.

12.4. Automatic Compensation of an Audio System Spectrum Operating with a Random Noise Input

CHARLES E. MAKI, MB Electronics, New Haven, Conn.

A set of 80 filters is used to divide a random noise input spectrum of an audio system into 25-cps increments. An identical set operating as a spectrum analyzer provides a similar function on the system output. By means of an automatic regulating system using solid-state electronics, each of the 80 loops is closed independently, thus providing a unique control sys-

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^{*} Sponsored by the Professional Group on Audio. To be published in Part 7 of the 1960 IRE INTERNA-TIONAL CONVENTION RECORD.
lags arise from the inductive properties of the excitation coil and the clutch rotor induced eddy currents. The servo compensation for the excitation coil time lag consists of excitation current feedback to decrease the effective transfer function time constant. Basic servo stability is provided by a tachometer feedback branch. Optimized compensation for the first-order time lag arising from the induced rotor eddy currents is provided by a passive tachometer lead network.

9.3. Synthesis of a Self-Adaptive Autopilot for a Large Elastic Booster

GEORGE W. SMITH, The Martin Co., Denver, Colo.

The design of an autopilot system for a highly elastic booster requires taking into consideration a vehicle with a flexible structure having relatively low bending frequencies and very little structural damping. If any of the systems parameters deviate from the calculated nomial values, the system stability margins become dangerously low. This results in the conventional autopilot design being dependent on a priori knowledge of the exact structural parameters. To alleviate this situation, the design of a self-adaptive autopilot has been evolved, which will automatically adjust itself so as to maintain optimum performance in the face of changing structural parameters. Because the system involves rigid body modes, as well as elastic modes, the system in general has a characteristic equation of high order.

An autopilot system is described which makes provision for continuous interrogation of the system to determine its response. Criteria are developed for adjustment of internal gains and filter parameters so as to achieve optimum response. Mechanization of the autopilot as a sampled data control system is discussed.

9.4. Design of Optimum Beam Flexural Damping in a Missile by Application of Root-Locus Techniques

R. J. HRUBY, Northrop Corp., Hawthorne, Calif.

The control of a missile in free-fall in the presence of external disturbances requires that the control system be stable when various body-bending modes are excited. The complexity of the control system depends upon the beam-flexural modal damping of the completed missile because all body-bending modes are excited by the rocket motor.

This paper explains how to use the subassemblies of a missile, such as the rocket motor or guidance package, to get a theoretical maximum increase in flexural damping by the proper design of the attaching fixtures of the subassemblies without weight or equipment complexity.

9.5. Flywheel Control of Space Vehicles

JAMES E. VAETH, The Martin Co., Baltimore, Md.

This paper will describe, and summarize the results of, an analytic study of inertia flywheels as employed for simultaneous three-axis attitude control of space vehicles. The accomplished objectives of this study were to determine the requirements, capabilities, and limitations of flywheel autopilots as functions of desired accuracy and speed of response, disturbing moments, differential gravity restoring torques, component uncertainties, and vehicle initial attitude and attitude rate errors. The performance degradation (from the viewpoint of stability and accuracy) that results from *not* compensating for gyroscopic cross-coupling torques was determined by means of a three-axis analog computer study.

SESSION 10*

Tues. 10:00 A.M.-12:30 P.M.

Waldorf-Astoria Astor Gallery

DIRECT CONVERSION

Chairman: WILLIAM A. HIGIN-BOTHAM, Brookhaven Natl. Lab., Upton, N. Y.

10.1. Thermoelectric Converters

KURT KATZ, Westinghouse Electric Corp., Pittsburgh, Pa.

Many advantages could be realized if it were possible to convert the heat energy produced, through fission, in a nuclear reactor directly to electrical energy. This paper will review the equations which define thermoelectric conversion efficiency, and the present status of materials for this application.

Several schemes for the conversion of fission-generated heat directly within the core zone will be described. The transfer of the fission-generated heat to an external thermoelectric converter unit will also be presented.

The advantages and disadvantages of each scheme will be described.

10.2. Thermionic Converters

WALTER GRATTIDGE, General Electric Res. Lab., Schenectady, N. Y.

A thermionic converter is a static hightemperature engine that converts heat to electricity. It is a vacuum or gas-filled device containing a hot cathode and a cold anode. Nuclear energy is an ideal heat source for a thermionic converter. The cathode and nuclear fuel element can be designed as an integral unit. The anode can be cooled by a conventional reactor coolant. Thus the converter can be an integral part of the fuel assembly. The anode temperature can be sufficiently high to permit the waste heat to be used in standard steam turbines. Thus the converter becomes a "topper" to the steam turbine and augments its performance.

10.3. Noble-Gas Plasma-Diode Thermionic Converter

F. E. JAMERSON, General Motors Res. Labs., Warren, Mich.

The direct conversion of nuclear heat to electricity is being investigated in a plasma diode that utilizes fission fragments from a uranium-bearing cathode to produce a plasma in a noble gas. The fractional ionization is relatively low, but the electron mobility in a noble gas is relatively high so that a reasonable diode impedance might be attained. The advantage of such a diode converter is that it eliminates the use of an alkali metal as used in other plasma diodes. The results of an experiment with a noble-gas plasma diode operated in the University of Michigan reactor will be discussed. In this experiment a uranium carbide cathode was operated up to a temperature of 2000°K. Some results of an initial theoretical analysis of the noble-gas plasma diode will be given.

10.4. Magnetohydrodynamic Approaches

RICHARD J. ROSA, Avco-Everett Res. Lab., Everett, Mass.

This paper starts with a brief review of the basic theory of magnetohydrodynamic (MHD) power generators and describes an experimental MHD generator which has been built at the Avco-Everett Research Laboratory.

MHD power plant cycles are discussed. A cycle designed for use with a nuclear reactor is described. The reactor in this case would need the same temperature capability as those now being developed for nuclear rocket propulsion although it might not, depending upon the application, require such a high specific power output.

The basic electrical properties of gases are reviewed briefly, and their dependence upon temperature, pressure, and composition is discussed. Over-all performance characteristics of MHD generators are presented as a function of these gas properties. This gives an indication of the future potential of MHD as a conversion technique.

Some of the possibilities for the future are discussed in which more exotic forms of hightemperature reactor combined with an MHD generator may make possible systems having, by present standards, very high efficiency, low cost, and/or light weight.

10.5. Direct Conversion—Where Do We Stand?

ROBERT J. PIDD, General Dynamics Corp., San Diego, Calif.

Several schemes have been advanced for the direct conversion of heat into electricity. These schemes will be reviewed and compared with respect to such technical parameters as efficiency and output and such programmatic factors as the state of the art and development required.

^{*} Sponsored by the Professional Group on Nuclear Science. To be published in Part 9 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SESSION 8*

2:30-5:00 P.M.

Mon.

Coliseum Morse Hall

ELECTRON DEVICES

Chairman: A. E. ANDERSON, Western Electric Co., Allentown, Pa.

8.1. Rating Power Transistors for High Current Pulses

PETER BALTHASAR, Bendix Aviation Corp., Long Branch, N. J.

In most switching applications the maxinium permissible collector junction temperature is the limiting factor. The junction temperature is a function of power, time, thermal resistances, and time constants. Therefore, the applied instantaneous power may exceed the rated dc power dissipation. In order to determine the equilibrium value of the peak junction temperature, theoretically derived equations are applied to the problem. Simplified approximations are made and compared to measurements. A procedure is outlined for determining by various design aids whether a transistor can be used for a certain pulse amplitude, pulse width, and pulse repetition rate combination.

8.2. An N-P-N Fusion Alloy Silicon Transistor for "Avalanche Mode" Operation

R. C. WONSON AND W. A. MCCARTHY, Raytheon Co., Newton, Mass.

This paper describes an n-p-n silicon transistor operated in the "avalanche mode." Normally the transistor has a very low frequency limitation; it has switching times of approximately 2 μ sec. When operated in the "avalanche mode," the transistor exhibits switching times of 2 nussec or better.

Since the transistor is silicon, its high temperature performance has been examined in detail. Measurements techniques are described as well as inexpensive methods of determining performance.

Practical circuit applications are discussed with particular emphasis on a completely designed pulse generator.

8.3. Photoconductor Optical Encoders for In-Line Readout Devices

CARL ISBORN, Beckman/Berkeley Div., Richmond, Calif.

This paper describes a method of using photoconductors to convert a four-line binary code into a seven-line code suitable for actuating a seven-segment visual display. The basic photoconductor amplifier is analyzed and performance data on typical mass-produced photoconductors are given. Photoconductor circuits that perform AND, OR, and NOT operations by converting electrical signals to optical signals and back to electrical signals are described. The logical equations defining the four-line to sevenline conversion are presented, and a conventional diode matrix for performing these logical operations is shown. An equivalent optical matrix is developed wherein light beams replace matrix diodes. The two conversion methods are compared with respect to performance, circuit complexity, and cost.

8.4. Advances in Screen Structure and Data Distribution for the ELF Display System

E. A. SACK, Westinghouse Res. Labs., Pittsburgh, Pa.

Microcircuit structures have been developed for the electro-luminescent-ferroelectric (ELF) display screen which achieve a display resolution of 256 elements per square inch. The switching matrix which distributes brightness information to the elements is an integral part of the screen. At present, multijunction silicon diode strips provide the switching function.

Model screens as large as eight inches wide by four inches high have been assembled. The highlight brightness of experimental models is on the order of 10 foot-lamberts with contrast ratios as high as 100/1. In one display system, information is distributed to the screen at a rate of 150 µsec per screen line with a frame-storage time of 5 seconds.

8.5. Shadow Grid VHF RF Tuner Tubes

F. R. SNYDER AND C. D. MCCOOL, General Electric Co., Owensboro, Ky.

The addition of a grounded grid aligned with the screen grid, and placed between the control grid and screen grid, will reduce partition noise as a result of the increased plate current to screen current ratio. A plate to screen current ratio of 60/1 or more is obtainable with this construction, which is termed the shadow grid technique.

A shadow grid pentode VHF-RF tuner tube has been developed having a noise figure 1.5 db better than standard tetrode RF tuner tubes, and has 6 db more gain.

The shadow grid technique permits the plate voltage to be higher, equal to, or lower than the screen voltage. This circuit flexibility aids the equipment designer significantly in obtaining cost savings.

8.6. Focus-Reflex Modulation for Electron Guns

KURT SCHLESINGER, General Electric Co., Syracuse, N. Y.

A new approach to high-efficiency guns for cathode-ray tubes has been developed. Emission from a large Pierce cathode is focused by deceleration, thus forming a small virtual cathode at the exit of a hyperbolic lens. Forward emission from this virtual cathode is controlled by an electron mirror before going through a spot-defining aperture. In operation, the mirror controls both the beam intensity and the focal length of the system. Ordinarily, both control functions are not coordinated.

In the present gun, fccus modulation and intensity modulation are brought into coincidence by design. The resulting "focus-reflex" modulation features total cutoff and high control sensitivity (10 to 12 volts) as well as high beam current (1200 μ a) and transconductance (G max = 300 μ a/volt). Spot size is under control by an aperture.

The paper will describe the electron-optical design theory for focus-reflex modulation.

Photographs of finished FR gun assemblies and of their characteristics will be shown. Television applications will be mentioned as well as an adaptation of the new gun to radar service.

SESSION 9*

Tues.

10:00 A.M.-12:30 P.M.

Waldorf-Astoria Starlight Roof

CONTROL APPLICATIONS

Chairman: HAROLD CHESTNUT, General Electric Co., Schenectady, N. Y.

9.1. Decoupling Techniques in Multiloop Control Systems

ROBERT H. LOOMIS, Westinghouse Electric Corp., Baltimore, Md.

Decoupling is a design technique which permits the independent design of the several transfer functions in a multiloop control system.

It is described in this paper by illustrating its application to a particular system and describing the results attained.

The technique described has the following advantages:

- The required "matching" takes place within the loop, so that the effect of any mismatch is attenuated by the open loop gain.
- For the example used, the output angular rate accuracy becomes independent of aircraft motion.
- Subject to the rather minor restriction implied by item 1, the two loops become independent as far as stability is concerned.

9.2. Optimum Compensation of a Position Servo with a Magnetic Clutch Actuator

R. J. HRUBY, Northrop Corp., Hawthorne, Calif.

This paper presents a derivation of the two first-order time lags typical of magnetic clutches used for control actuators in missile servo applications and an exact analysis of two specific methods used for servo compensation of the magnetic clutch time lags. The two time

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^{*} Sponsored by the Professional Group on Antomatic Control. To be published in Part 4 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

6.3. The Nature of Astro Doppler Velocity Measurement

JOHN E. ABATE, Kearfott Co., Inc., Clifton, N. J.

Astro Doppler velocity measurement for space vehicle navigation requires the use of electro-optical systems capable of measuring a small incremental change in the wavelength of propagated stellar energy. Such systems prowide velocity data whose character and limitations are a function of the star's radiation as well as the system instrumentation. This paper discusses the velocity data, their character and limitations; and the Doppler velocity systems, their capabilities, limitations, and instrumentation.

6.4. Generation of Artificial Electronic Displays, with Application to Integrated Flight Instrumentation

G. H. BALDING, Kaiser Industries, Palo Alto, Calif., AND C. SÜSSKIND, University of California, Berkeley, Calif.

A novel method is described by which complex electronic signals are generated, dynamically varied, mixed, and presented on a standard video (raster-scan) display, without the use of vidicon or similar camera devices. The method is applied to the generation of a "contact-analog" display that provides a stylized representation of the real world to the pilot of an aircraft. This integrated display, which varies continuously as the speed, altitude, or attitude of the aircraft changes, also contains a "flight-path" representation that enables the pilot to maintain a prescribed path.

6.5. The Synchro-Magnetic Approach and Terminal Landing System for Aircraft

Ross Gunn, The American University, Washington, D. C.

A reliable system for the guidance and control of aircraft near airports, throughout the period of approach, landing, and taxiing, has been developed.

The system employs a guide cable, excited by alternating current, which is laid, buried, or carried on poles along the projection of a preselected straight or curved flight path. The magnetic field pattern established by the current is sensed on the aircraft as it flies more or less parallel to the guide cable up to altitudes of more than 2000 feet. The magnetic pattern permits the simultaneous and unambiguous determination of the approximate altitude, the horizontal position of the aircraft with respect to the guide cable, the angle of yaw, and the direction of the flight along the cable. The response of all elements increases regularly as the touchdown point is approached, so that precision control of the flight path by automatic devices is entirely practicable.

Equipment is available that works successfully anywhere within 2000 feet of the guide cable. Greater ranges are accessible. Basic principles and detailed measurements of performance on a one-tenth scale landing range will be presented. The new system has many valuable operating features including simplicity and great reliability. **SESSION 7***

Mon.

2:30-5:00 P.M.

Coliseum Marconi Hall

PRODUCTION TECHNIQUES

Chairman: ROBERT L. SWIGGETT, Photocircuits Corp., Glen Cove, N. Y.

7.1. Fabrication and Interconnection of Microcircuits Applicable to Data Processing Equipment

J. E. RICHARDSON AND J. W. BURKIG, Hughes Aircraft Co., Culver City, Calif.

The feasibility of microminiaturizing data processing equipment is discussed. The following topics are covered:

- The fabrication of circuit elements with attention to controlling parameters to predetermined values within given tolerances.
- The problem of interconnection of individual circuit elements into circuits and the interconnection of aggregates of circuits into assemblies.
- 3) Problems of heat density, maintainability and reliability.

The discussion of application to data processing equipment is intended as a specific example. The discussion is equally applicable to the general field of microminiaturizing electronic equipment.

7.2. Ultrasonic Welding of Electronic Components

W. C. POTTHOFF, C. F. DEPRISCO, AND W. N. ROSENBERG, Aeroprojects Inc., West Chester, Pa.

A brief review of the ultrasonic welding process is given, with brief description of spottype, continuous-seam, and ring-type welding equipment of interest to the electronics industry. Specific applications of ultrasonic welding to fabrication and encapsulation of transistors, transformer coils, bridgewire assemblies, electronic tubes, etc., are discussed.

The absence of electric current, thus eliminating arcing, sparking, and outgassing to contaminate surrounding areas; the low temperatures induced during welding; the fact that ultrasonic welds are solid-state bonds; the absence of intermetallics in the weld zone; and the elimination of complicated wet chemical prewelding operations, constitute advantages in processing and product quality that recommend consideration of this unique joining process for electronic components.

7.3. A Disquisition of the Innovations and Gadgetry Used in the Volume Production of a Super Power Electron Device

JAMES A. JOLLY, Eitel-McCullough, Inc., San Carlos, Calif.

The X626 is a super power klystron of gigantic size. It was specifically developed to produce over 80 kw of average power (1.25 mw peak power) at UAF frequencies for the Air Force Project BMEWS (Ballistic Missiles Early Warning System). Many papers have been written describing the complexity of producing microminiature electron devices. This paper is unique in that it describes the innovations and novel applications of production methods, procedures, techniques, and processes necessary to volume produce an electron device weighing more than 700 pounds and standing over 10 feet tall. A large number of detailed photographs are used to describe accurately the manufacturing operations.

7.4. Design and Manufacturing of a Simplified Grid Module

LEON JACOBSON, General Electric Co., Syracuse, N. Y.

This paper deals with a simplified method of assembling and connecting commercially available components in encapsulated modules.

Eyelets positioned on a perforated pallet are used to make the mechanical and electrical connection between the component leads and the interconnecting wires that have been inserted in these eyelets. These eyelets are dipsoldered while the assembly is still on the pallet.

The resulting web of components, wire, and eyelets is removed from the pallet and then folded or rolled, before encapsulating, to produce the completed module.

Having the layout and assembly in one plane simplifies assembly and also predetermines the position of the components.

7.5. Micromodule Components: A Review of the State of the Art

R. A. FELMLY, RCA, Somerville, N. J.

Component parts which perform the basic electronic-circuit functions have been developed for use with the micromodule system. Versatility of these parts has been demonstrated by a variety of modules designed for use in the types AN/PRC-36 Helmet Radio and the AN/TCC-26 Time Division Multiplex Equipment. This paper describes the mechanical characteristics and construction, electrical specifications, and performance of these various types of components.

Among the components now in microelement form are precision-film resistors, single and multilayer ceramic capacitors in both high-K and NPO bodies, tantalum electrolytic capacitors, germanium transistors, gold-bonded germanium and diffused silicon diodes, and microminiature toroidal coils for inductor and transformer applications. The latter components include IF, RF, and pulse types.

Components under current development include a trimmer capacitor, thermistors, RC filters and extended-range resistors, capacitors and inductors. All of these components are being designed to the micromodule form factor.

^{*} Sponsored by the Professional Group on Production Techniques. To be published in Part 6 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

5.1. Management and the Employee-Owned Concept of Young R and D Growth Firms

D. M. KRUCHKO, Aero Geo Astro Corp., Alexandria, Va.

This paper presents a discussion of the problems and challenges of relatively new electronic R and D firms where relatively youthful technical personnel are subject to a new way of life surrounded by overhead cost distribution, operating costs, amortizations, depreciations, net worth, etc. It describes the unique features of an employee-owned firm where over 90 per cent of the employees are also stockholders.

The rewards, tangible and intangible, of being entrepreneurs in a society of corporate giants are detailed. In addition, a presentation is made of the advantage of flexibility and individual enthusiasm growing out of the employee-owner concept.

The paper concludes with a detailed presentation of the advantages the small R and D firms offer customers, both Government and commercial.

5.2. An Engineering Management View of the Maintainability Problem

M. J. MARCUS, IBM Corp., Owego, N. Y.

Management can take an important hand in the increase of product maintainability by recognizing the management factors in development engineering which bear directly on the eventual maintenance problems of the customer.

The maintainability of a system is usually evaluated only after the system has been placed in operation. It is then that customer and contractor maintenance personnel point out maintainability design deficiencies and expensive retrofit begins. This feedback loop must be shortened by detailing the implications of design decisions for system maintainability at the time development decisions are made.

Management techniques for insuring this shortened feedback loop on maintainability are described and discussed.

5.3. Engineering Management for Creative Appraisal of New Ideas— The Secret Weapon for Technical Progress?

WILLIAM H. BENUBIEN, General Electric Co., Utica, N. Y.

Effective appraisal of ideas is probably more important than conceiving them. It is very challenging in engineering management to exercise techniques for idea appraisals that can effectively select and promote the best ones without mwittingly eliminating some that may later emerge from other competitive sources with a tremendous scientific or technical impact. This usually results from man's inherent tendency to look for what is wrong rather than assuming an attitude of "How can we make it work?"

Planned engineering management programs employing creative idea appraisal techniques are shown to result in dramatic increases in new products, technologies, services, and operating practices. The widely varying degree of technological impact that has resulted indicates the universal nature of the application possibilities of this new management tool.

5.4. How to Produce Reliable Products at a Profit

C. W. WATT, Raytheon Co., Waltham, Mass.

Active and rapid feedback channels from detecting points to action points in the organization are essential if an over-all manufacturing-engineering system is to produce reliable and profit-making products. That the concept of negative feedback is a valid one organizationally can be seen by observing its operation in manufacturing, where it is realized as the quality control function. A similar monitoring function should exist formally in engineering; and when it does it can lead to better performance of the engineering job. One way to establish such a function is through the medium of the design review, which can be effected in most cases with no increase of personnel. As a corollary, the specialists in the engineering organization can be most effectively used when they, taking a team approach, can provide a balanced consulting service to the designers with whom they work, and also participate actively in design reviews.

5.5. Concepts of Capital Financing for Electronic Companies

R. T. SILBERMAN, Electronics Capital Corp., San Diego, Calif.

During the last decade the nearly ten-fold expansion of the electronics industry has brought with it a myriad of different approaches to the solution of working capital problems. These principally have included private and public equity financing, Government "V" loan credit, short-tern bank credit, and accounts receivable financing and vendor financing in the form of delayed settlement of accounts payable. Internal cash flow naturally has been generated from depreciation and profits—where, in fact, the operations have been successfully managed.

It is now possible to look back and bring into proper focus the consequences of each method of financing and how these methods have affected the company growth, market position, and the capital appreciation of the company founders and stockholders. The lessons of our past financial activities define objectives for future financing of existing and new electronic enterprises. The heavy hand of direct Government contracting has specific influence in determining the optimum financial plan. The prospect of continuing inflation, high interest rates, and a generally controlled economic environment make certain forms of debt financing most attractive when in proper balance to earned surplus and equity.

Special legislative action during the past years in the creation of Small Business Investment Companies provides another vehicle for capital expansion in the next decade. Factors for determining the true cost of financing when balanced against growth rate, market appreciation, and owner equity are discussed.

SESSION 6*

Mon.

2:30-5:00 P.M.

Coliseum Faraday Hall

ADVANCES IN AEROSPACE SUBSYSTEMS

Chairman: D. G. C. LUCK, RCA, Princeton, N. J.

6.1. Range Ambiguity Resolution in High PRF Radar

NORMAN S. POTTER, The W. L. Maxson Corp., New York, N. Y.

The origins of high pulse recurrence rates in radar and the fundamental means of resolving the resultant ambiguities are quantitatively discussed. An analytical study is conducted of the information requirements for the determination of target range as opposed to detection alone. A statistical model of the detection and ambiguity resolution process is constructed for several generalized radar-computer system configurations. A new general theory is developed and applied to determine quantitatively the relationship between system memory and range resolution for information handling schemes that bracket all ambiguity resolution techniques.

For purposes of demonstration, the theory is applied to range quantized systems, and appropriate digital techniques are developed for the processing of multiple PRF coded data. The required radar signal modulation is functionally related to operational and system parameters. Eclipsing losses and the degradation of the efficiency of the range ambiguity resolution subprogram due to the simultaneous presence of multiple objects in the beamwidth are statistically investigated and signal modulation methods that minimize these problems are developed.

6.2. An Ion Altimeter for Pressure-Altitude Measurements

G. V. ZITO, Bendix Aviation Corp., Teterboro, N. J.

A ruggedized, cold cathode radiation-type ionization air density gauge for pressure-altitude measurements is described together with the front-end circuitry required for instrumentation. Typical applications include the static pressure sensor channel of hypervelocity machmeters and altimeters where continuous indication is desirable and in air data computer systems for military and commercial aircraft. The hysteresis and position sensitivity of diaphragm-type sensors are largely absent in this class of device. Historical background on the development of extreme altitude sensors, and pertinent data on calibration procedures and sensor compensation are also included.

^{*} Sponsored by the Professional Group on Aeronantical and Navigational Electronics, To be published in Part 8 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SESSION 4*

Mon.

2:30-5:00 P.M.

Waldorf-Astoria Sert Room

RADIO FREQUENCY INTERFERENCE

Chairman: RALPH M. SHOWERS, University of Pennsylvania, Philadelphia, Pa.

4.1. Simulation Tests on an Interference Rejection Antenna System

W. D. WHITE AND C. O. BALL, Airborne Instruments Lab., Mineola, N. Y.

The antenna investigated consists of three identical receptor elements spaced in a suitable configuration. They are so arranged that at the output terminals the desired signal can be represented by three identical phasors. Due to the physical separation however, an interference signal will be represented by three phasors having identical lengths but different phase angles. The resultant of signal plus interference will thus produce three phasors lying on the circumference of a circle. A computer which is capable of finding the center of this circle will then deliver an output signal uncontaminated by the interference.

The performance of the system is limited by the accuracy with which the three receptor elements can be made identical and by the effects of receiver noise. To investigate the latter limitation, a digital computation program was set up to simulate the operation of the system. The results indicate that if advantage is taken of the fact that the relative jammer phases are likely to be slowly varying, then, with reasonable separation of the interference phasor, the residual interference signal can be reduced below the level of receiver noise.

4.2. Computer Simulation of Signal Environments

W. G. JAMES, American Machine & Foundry Co., Alexandria, Va.

As an aid to designers, a digital computer simulation program which can determine the signal environment resulting from a large deployment of electromagnetic emitters has been developed for use with the IBM 704 computer. The program provides realistic simulation of such emitter characteristics as antenna gain patterns and scanning cycles and of such propagation phenomena as atmospheric absorption and tropospheric refraction. A deployment of some 5000 emitters including a variety of some **100** types has been simulated. The program design permits extensive expansion of the number of emitters and emitter types in the simulated environment.

4.3. Wiring of Data Systems for Minimum Noise

J. V. WHITE, Beckman/Systems Div., Anaheim, Calif.

Noise sources due to undesired coupling of circuits in a data system can be minimized by proper wiring. Construction of a system central ground, which avoids ground-resistance coupling, is described. To eliminate the remaining common-resistance coupling, this construction is extended to include all system nodes. Proper location of the ground and nodes permits wiring all noise-sensitive circuits with twisted cable, minimizing coupling through mutual inductance. Proper shielding for protection from electrostatic interference is simplified. A set of explicit wiring rules, which implement this method, is given. Examples of the application of these rules are included.

4.4. Receiver Analysis for Interference Prediction Purposes

D. C. PORTS AND C. R. MILLER, Jansky & Bailey, Washington, D. C.; AND J. SAVAGE, Rome Air Dev. Center, Griffiss .1 FB, N. Y.

A method will be presented to describe a means of evaluating receiver behavior for a degree of susceptibility to various types of undesired signals. To determine acceptable levels of RF input energy, it is necessary first to examine signals at the receiver output terminals to establish a required level and quality of desired intelligence. Examination proceeds from output to RF input on a function-by-function basis to determine the modifying effects caused by pertinent characteristics of RF, IF, and detector circuits, on the intelligence as it passes through the receiver. Results are combined to produce an expression for a transfer function. Intelligibility adjustments are obtained from this to deduce the degree of control exerted by the portions of the receiver, and this enables the development of a technique for prediction of level of interference to be anticipated.

4.5. Electromagnetic Interference and Vulnerability Reduction

JOHN J. EGLI, U. S. Army Signal R & D Lab., Fort Monmouth, N. J.

Interference and vulnerability, the first signifying a mutual or unintentional interference, and the second signifying an intentional interference, are omnipresent disturbances which must be reduced to permit tactics of the military to be consummated without extraneous impairment.

In the research and development area of electromagnetic interference and vulnerability reduction, three important phases are essential to achieve reduction: 1) theoretical analysis, 2) design criteria, and 3) instrumentation and measurements. A research and development organization capable of handling these three important phases has been established at the U.S. Army Signal Research and Development Laboratory, Fort Monmouth, N. J. While environment testing is a require-

While environment testing is a requirement for a well-integrated reduction program and will be provided as a facility at the U. **S.** Army Electronic Proving Ground, Fort Huachuca, Ariz., this facility will not be a subject for discussion in this paper.

4.6. Fire Fighting or Fire Prevention

L. A. YARBROUGH AND J. W. WORTHINGTON, JR., Hdqrs. GEEIA, Griffiss AFB, N. Y.

The subject of reduction or elimination of interference to Communications-Electronics equipment, facilities, or systems has been the topic of many papers, conferences, meetings symposia, and other get-togethers. Many techniques, in theory, are known to reduce or eliminate interference. Unfortunately, not many "black boxes" are available to implement the theoretical solutions.

This paper will present the aggressive twofold program to prevent or reduce interference being implemented by the USAF Ground Electronics Engineering Installation Agency (GEEIA). Methods of fire prevention discussed will include advance planning and adequate pre-engineering effort, proper siting, equipment or system interference specifications, and installation techniques. Details of GEEIA's "fire fighting" capability, consisting of contractor furnished teams, plus an "in-house" capability will be discussed. Future plans to reduce or prevent interference to USAF CE systems will be discussed.

Mon.

2:30-5:00 P.M.

Waldorf-Astoria Empire Room

SESSION 5*

ENGINEERING MANAGE-MENT-I

Chairman: H. M. O'BRYAN, Sylvania Electric Products, Inc., New York, N. Y.

^{*} Sponsored by the Professional Group on Radio Frequency Interference. To be published in Part 8 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

^{*} Sponsored by the Professional Group on Engineering Management. To be published in Part 10 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

2.2. Transistorized Radiation Monitoring Equipment

J. J. HENRY, Union Carbide Nuclear Co., Oak Ridge, Tenn.

An integrated monitoring system is described which features modular hermetically sealed construction and which operates from 115-volts, 60-cycle ac power. It consists of a completely transistorized linear preamplifier, amplifier, discriminator, and count rate meter. The amplifier is extremely stable and has a band pass greater than 5 megacycles, linearity of 0.5 per cent and excellent overload characteristics. The discriminator is nonsaturating and the rate meter features a diode switch, constant current, metering circuit. The entire system is capable of operating linearly at uniform repetition rates up to 500 kc. A variable low impedance output is provided for continuous potentiometric recording.

2.3. A Sensitive Parametric Modulator for DC Measurements

R. R. HOGE, Bendix Aviation Corp., Detroit, Mich.

This paper concerns a high-impedance semiconductor modulator for amplification of signal currents at levels which previously have required the use of electrometer tubes. An input threshold of 5×10^{-14} watts at 10^{-12} amperes is observed, and substantial signal power gain is provided by the modulator. The modulator is basically an ac Wheatstone bridge in which two arms are silicon capacitor diodes. Input signals modify the capacitance of the depletion region in the diodes and control the (un)balance of the Wheatstone bridge, Initial balancing adjustments are simple and permanent, and are limited to one or two components. Typical applications in transistorized nuclear instrumentation are indicated.

2.4. Semiconductor Synchronous Clamp for Millivolt Signal Levels

A. J. KOLL, E. BLECKNER, AND O. C. SRYGLEY, *Aero Geo Astro*

Corp., Alexandria, Va.

Semiconductor clamp and detecting circuits of the conventional type are subject to serious inaccuracies when applied in the millivolt range. This nonlinearity in operation is caused by the barrier potential which is inherent in semiconductor devices.

The purpose of the circuit described herein is to provide a method of accurately clamping pulses in the millivolt range to any desired dc level. It is also useful for detecting pulse amplitudes in the millivolt level and for storing this detected amplitude in a memory or learning time constant.

The basic circuit employs a bridge with two torward biased diodes and two Zener diodes. The diodes are arranged so that a voltage pulse applied to the bridge provides high conductivity to pass the low-level signal. When the pulse is removed from the bridge, the Zener characteristic prevents leakage back through the bridge element. This circuit operates well with low millivolt pulses over an extremely wide temperature range and it successfully meets a wide variety of extreme environmental conditions.

SESSION 3*

2:30-5:00 P.M.

Mon.

Waldorf-Astoria Iade Room

THE ENGINEER WRITES AND SPEAKS

Chairman: CHESTER W. SALL, RCA, Camden, N. J.

3.1. How to Edit Your Own Papers

ELEANOR M. MCELWEE, RCA, Harrison, N. J.

This paper proposes to begin where most "how-to-write" articles end, *i.e.*, to show how engineers can improve their reports, papers, and proposals after they write them. Editing is shown to be simply a critical appraisal of written material to make sure that it says exactly what the writer intends and will not be misinterpreted. Several common writing faults which tend to obscure meaning are analyzed, and a "do-it-yourself" approach is described for recognizing and correcting them in a preliminary draft. The topics covered include organization of material, testing for relevancy and sufficiency by means of an outline, paragraphing, function of subheads, a diagrammatic method of sentence editing, the logical basis for rules of grammar and punctuation, use of abbreviations and symbols, and handling of forumulas, tables, and illustrations. A "check list of major items is included for use by engineers in the editing of their own papers, reports, or proposals.

3.2. Basic Concepts of Increased Effectiveness in Oral Presentations

IRVING J. FONG, Remington Rand Univac, Div. of Sperry Rand Corp., St. Paul, Minn.

The danger of our present system of information documentation and retrieval collasping under its own weight has been partly alleviated by the creation of auto-abstracts, produced entirely by machine techniques. Computers now can read from a document directly, and with machine-generated thought processes, produce abstracts and indexes to save reader-time and effort.

The human creator, in teaching the machine to simulate his own patterns of learning and thought processes, in the field of scientific communication has neglected his own natural medium of communication to an audience, that is, the spoken language or oral presentation. A return to basic concepts that could increase the effectiveness of verbal presentations of technical papers is given. This is intended to balance the progress made by recent machines in producing abstracts, since an oral presentation is in effect an amplified abstract, whose purpose also is to facilitate the selection and utilization of useful information.

3.3. New Horizons in Scientific Information Preparation

N. J. SMITH, IBM Corp., Kingston, N. Y.

The overwhelming volume of scientific documentation has almost succeeded in making the retrieval of information an impossibility. Three International Conferences on Scientific Information have been held to define problems of information retrieval and to investigate automated methods of solution.

As a result, the next phase of study will surely be a reappraisal of the methods and techniques of preparing scientific reports and documentation. The present method of composing reports and documentation by writing is essentially only a mode of information storage. In the final analysis, the efficiency of retrieval is determined by the method of information preparation, storage, and transfer. And since the optimum solution of the retrieval problem will lead inevitably to a study of the information preparation and storage, it is, indeed, the time for Engineer Writers to examine the basic concepts of language as a science and not as an art. New formats for idea arrangement and connotation control are even now being developed. These formats lend themselves to automated techniques of storage and, hence, to automated research and retrieval.

This paper discusses the status of automated retrieval systems and the problems and possibilities of the new techniques of information preparation. It describes the disciplines involved and discusses the present state of the art. It suggests areas of investigation for those who must be ready for these new requirements for the preparation of scientific information.

3.4. The Paper Reader at Conventions Will Soon Be Obsolete!

J. O. REECE, Texas Instruments Inc., Dallas, Tex.

For many years the trend of authors presenting technical papers at conventions and symposiums has been to read the paper from beginning to end. At recent technical conventions, however, the paper reader has been outlawed.

Most speakers at conventions read their speeches in such a deadly monotone that it actually hulls the listeners to sleep. The reader of a speech seems to lose all contact with the audience. The paper reader somehow fails to realize that he owes a responsibility to the audience. That responsibility is to make his speech interesting to everyone in the room so that the entire audience wants to hear what he is saying instead of having the feeling of being forced to listen to him.

This paper will describe why technical conventions in the near future will be operating by a new format, that of an orally presented paper, and why the paper reader will soon be obsolete.

World Radio History

^{*} Sponsored by the Professional Group on Engineering Writing and Speech. To be published in Part 10 of the 1960 IRE INTERNATIONAL CONVENTION REC-ORD.

ABSTRACTS OF TECHNICAL PAPERS

SESSION 1*

Mon.

2:30-5:00 P.M.

Waldorf-Astoria Starlight Roof

CONTROL THEORY

Chairman: JOHN C. LOZIER, Bell Telephone Labs., Inc., Whippany, N. J.

1.1. Incremental Phase Plane Analysis of Nonlinear Sampled Systems

J. A. ASELTINE AND R. A. NESBIT, Space Technology Labs., Inc., Los Angeles, Calif.

An analysis technique for nonlinear sampled data systems is developed, using the incremental phase plane. This method is analogous to the phase-plane method for continuous systems. A sampled data system with saturation is analyzed to demonstrate the use of the incremental phase plane in system analysis. Path tangent curves are introduced which allow the graphical solution of systems with general types of nonlinearities. A difference equation analog of van der Pol's equation is solved using the path tangent curves.

1.2. On the Existence and Uniqueness of the Optimal Multivariable Systems Synthesis

MIHAJLO D. MESAROVIĆ, Case Inst. Tech., Cleveland, Ohio

The possibilities of achieving a totally optimal performance in multivariable systems have been considered for filter and control-type problems. For a functional performance criterion and a linear system, the derived optimizing equations do not uniquely define an optimal system. Additional specifications are necessary for a complete determination of the system. The various types of synthesis could be characterized by these additional specifications. For nonlinear systems in general the derived equations, based on performance criteria, uniquely define the optimal system. Nonlinear multivariable systems could be classified according to completeness in determining an optimal system using a certain type of performance criterion. For a set of random inputs the performance criteria are expressed as a mean over some function of the outputs. In such a case minimizing equations cannot in general be satisfied by the characteristic functions of the system to be synthesized. The relatively optimal systems can be, however, uniquely defined.

1.3. On Optimal and Suboptimal Policies in the Choice of Control Forces for Final-Value Systems

MASANAO AOKI, Dept. of Engrg., University of California, Los Angeles, Calif.

Some discrete final-value systems are formulated as stochastic multistage decision processes.

A functional equation is set up and solved to obtain an optimal sequence of control forces (henceforth referred to as an optimal policy) for a given performance index.

The procedure is illustrated for the case of the usual minimization of the mean-square of the final error, with discussions of analytic and computational techniques.

A simpler yet near-optimal policy is conjectured and performances of final-value systems with the optimal and the suboptimal policies are compared by the Monte Carlo method.

An approximate analysis of the system behavior resulting from the suboptimal policy is also presented.

1.4. A Study of Asynchronously Excited Oscillations in Nonlinear Control Systems

O. I. ELGERD, Dept. of Elec. Engrg., University of Florida, Gainesville, Fla.

This study was initiated as an effort to explain certain oscillatory phenomena observed in an aircraft and weapon control system operated in a particular mode. The oscillations in question were of the undamped limit-cycle type and their presence could very clearly be correlated with the degree of noise corruption of the signals.

It is demonstrated in this paper that certain types of nonlinear systems, although inherently stable, may be driven into oscillatory modes not only by random signals but also by any high-frequency periodic or nonperiodic signal possessing a certain energy content. It is of interest to note that the tendency for hunting and also the hunt frequency usually are completely independent of the frequency of the excitation signal, *i.e.*, the phenomenon is of asynchronous nature.

A general theory, the validity of which has been tested by both analog and digital means, is presented and utilized to demonstrate how this phenomenon may be predicted from information on circuit data.

1.5. On the Optimum Synthesis of Sampled Data Multipole Filters with Random and Nonrandom Inputs

II. S. HSIEH AND C. T. LEONDES, Dept. of Engrg., University of California, Los Angeles, Calif.

This paper considers the synthesis of optimum sampled data multipole filters with n inputs and m outputs. The signal portion of the input will consist of a stationary random

component and a nonrandom component which is assumed to be polynomial with unknown coefficients but known maximum order. Each signal is corrupted by stationary random noise. The filter under investigation is linear, time invariant, and with finite memory.

The design criterion is to specify the weighting functions of the filter so that the system error, which is defined as the difference between the actual and ideal outputs, has zero ensemble mean and the system ensemble mean square error is minimum. The weighting functions thus obtained will have, in general, abrupt jumps at the sampling instants, but they are continuous within the sampling intervals.

SESSION 2*

Mon.

2:30-5:00 P.M.

Waldorf-Astoria Astor Gallery

THE BROOKHAVEN ALTER-NATING-GRADIENT SYNCHRO-TRON; TRANSISTORIZED NUCLEAR INSTRUMEN-TATION

Chairman: RICHARD F. SHEA, General Electric Co., Schenectady, N. Y.

2.1. The Brookhaven Alternating-Gradient Synchrotron

Part I—Alternating Gradient Synchroton, J. P. BLEWETT, Brookhaven Natl. Lab., Upton, N. Y.

- Part II—The Linear Accelerator Injector for the AGS, S. D. GIOR-DANO, Brookhaven Natl. Lab., Upton, N. Y.
- Part III—Radio-Frequency Accelerating System for the AGS, M. PLOTKIN, Brookhaven Natl. Lab., Upton, N. Y.

The alternating-gradient synchrotron at Brookhaven National Laboratory is being prepared for initial operating. The 30-billion electron volts, \$30-million proton accelerator will be the first American machine to reach this energy. These three papers present some of the principles of design and the details of construction of the project.

^{*} Sponsored by the Professional Group on Automatic Control. To be published in Part 4 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

^{*} Sponsored by the Professional Group on Nuclear Science. To be published in Part 9 of the 1960 IRE INTERNATIONAL CONVENTION RECORD.

SCHEDULE OF TECHNICAL SESSIONS

	WALDORF-ASTORIA HOTEL			N	EW YORK COLISEU	M			
	Starlight Roof	Astor Gallery	Jade Room	Sert Room	Empire Room	Grand Ballroom	Faraday Hall	Marconi Hall	Morse Hall
Monday March 21 2:30 P.M 5:00 P.M.	Session 1 CONTROL THEORY	Session 2 THE BROOKHA- VEN ALTERNAT- ING-GRADIENT SYNCHROTRON; TRANSISTORIZED NUCLEAR IN- STRUMENTATION	Session 3 THE ENGINEER WRITES AND SPEAKS	Session 4 RADIO FREQUENCY INTERFERENCE	Session 5 ENGINEERING MANAGEMENT —1		Session 6 ADVANCES IN AEROSPACE SUBSYSTEMS	Session 7 PRODUCTION TECHNIQUES	Session & ELECTRONIC DEVICES
Tuesday March 22 10:00 A.M 12:30 P.M.	Session 9 CONTROL APPLICATIONS	Session 10 DIRECT CONVERSION	Session 11 BROADCASTING —1	Session 12 AUDIO		Session 13* ENGINEERING MANAGEMENT —II	Session 14 VARIED VIEWS OF MEDICAL ELECTRONICS	Session 15 MODERN AP- PROACHES FOR IMPROVED AIR TRAFFIC MAN- AGEMENT	Session 16 BROADENING DEVICE HORIZONS
Tuesday March 22 2:30 P.M 5:00 P.M.	Session 17 RADAR AND CODING THEORY	Session 18 INDUSTRIAL ELECTRONIC IN- STRUMENTATION	Session 19 BROADCASTING —11	Session 20 AUDIO AND BROADCAST AND TELEVISION RECEIVERS			Session 21 THE HUMAN AS ORIGINATOR OF SIGNALS AND SCHEMES	Session 22 DESIGN OF EQUIPMENT RELIABILITY	Session 23 MICROWAVE TUBES
Tuesday March 22 8:00 P.M 10:30 P.M.						Session 24 Panel: ELEC- TRONICS-OUT OF THIS WORLD			
Wednesday March 23 10:00 A.M 12:30 P.M.	Session 25 DETECTION THEORY AND APPLICATIONS TO PHYSICS	Session 26 BROADCAST AND TELEVISION RECEIVERS	Session 27 ELECTRONIC COMIONENT PARTS	Session 28 SPACE TELEMETRY		Session 29* SEMINAR ON 1959 ITU GENEVA CONFERENCE	Session 30 COMMUNICA- TION SYSTEMS DESIGN	Session 31 ASPECTS OF COMPONENT RELIABILITY	Sestion 32 MICROWAVE FILTERS
Wednesday March 23 2:30 P.M 5:00 P.M.	Session 33 ELECTRONIC COMPUTERS AND CIRCUIT THEORY: HOW EACH TECHNOLOGY CAN HELP THE OTHER	Session 34 ULTRASONICS ENGINEERING—I	Session 35 COMPONENT PARTS	Session 36 STEREOPHONIC SOUND REPRODUCTION			Session 37 COMMUNICA- TION SYSTEM TECHNIQUES	Session 38 ANTENNA PATTERN SYNTHESIS	Session 39 MICROWAVIE INTERACTION WITH MATTER
Thursday March 24 10:00 A.M 12:30 P.M.	Session 40 ADAPTIVE NETWORKS	Session 41 CIRCUIT THEORY: CURRENT CON- TRIBUTIONS	Session 42 ULTRASONICS ENGINEERING —H	Session 43 EQUIPMENT AND SYSTEMS	Session 44 SATELLITE COM- MUNICATIONS		Session 45 HUMAN FACTORS IN ELECTRONICS	Session 46 SCANNING ANTENNA ARRAYS	Session 47 MAGNETIC RECORDING
Thursday March 24 2:30 P.M 5:00 P.M.	Session 48 ELECTRONIC COMPUTERS	Session 49 SYMPOSIUM ON A DECADE OF PROGRESS IN NETWORK THEORY	Session 50 SPACE ELECTRONICS	Session 51 CHECK-OUT IN- STRUMENTATION AND CIRCUITRY			Session 52 VEHICULAR COMMUNICA- TIONS	Session 53 ANTENNA AND PROPAGATION PROBLEMS	Session 54 WAVEFORM ANALYSIS AND RANDOM VIBRA- TION

* Sessions terminate at 12:00 Noon.

1960 IRE INTERNATIONAL CONVENTION PROGRAM Waldorf-Astoria Hotel, New York Coliseum, March 21-24, New York, N.Y.

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CONVENTION HIGHLIGHTS

Technical Program

A schedule of 54 technical sessions appears on the next page, followed by abstracts of the 258 papers to be presented.

Radio Engineering Show

This year's exhibition will again be held in the New York Coliseum at 59th St. and 8th Ave. A list of the 850 exhibitors and their products appears in "Whom and What to See at the Radio Engineering Show" in the advertising section of this issue.

Annual Meeting

Time: 10:30 A.M., Monday, March 21. Place: Grand Ballroom, Waldorf-Astoria Hotel.

Speaker: Llovd V. Berkner, President of the Associated Universities, Inc., "Can the Social Sciences Be Made Exact?

The special features of this opening meeting of the convention will be of particular interest to all IRE members.

Annual IRE Banquet

Time: 6:45 P.M., Wednesday, March 23. Place: Grand Ballroom, Waldorf-Astoria Hotel.

The Annual IRE Banquet is in many ways the climax, not only of the convention, but of the entire year. It is at this time that the leading contributors to the progress of our profession are annually singled out for recognition by the IRE. An outstanding program of nationally prominent guest speakers and IRE officers will include the presentation of six awards and recognition of the 76 newly elected Fellows of the IRE.

Seats are reserved on a "first come, first served" basis. Place your order now.

Cocktail Party

Time: 5:30-7:30 p.m., Monday, March 21. Place: Grand Ballroom, Waldorf-Astoria Hotel.

Women's Program

An entertaining program of tours and shows has been arranged for the wives of members. Women's headquarters will be located in the Regency Suite on the fourth floor of the Waldorf.

World Radio History



1960

F. T. PIERCE

ematics and in 1938 the B.S. degree in physics, both with first class honors, from the University of Wales. He received the Ph.D. degree from the University of Cambridge in 1950.

During the war Dr. Pierce held posts under the British Ministry of Supply

and was particularly concerned with the development of aircraft armament. In 1946, he joined the faculty of the University of Cambridge, attached to the Solar Physics Observatory, where he did research work in the fields of atmospheric electricity, solar physics and the airglow. From 1950 to 1957, he was a University Teaching Officer at the Cavendish Laboratory, Cambridge, and was directly responsible for the research in the Meteorological Physics Section. This included work in atmospheric electricity. cloud physics, atmospherics and radio wave propagation, etc. From 1957 through 1958, he worked upon problems in atmospheric physics, gas discharges, and high voltage effects at Vickers Research Ltd., Weybridge, England, where he was Senior Scientist. Since January 1959, Dr. Pierce has been at AVCO RAD Division, Wilmington, Mass., as a Senior Staff Scientist, primarily dealing with topics in space and atmospheric physics.

Dr. Pierce is a member of Commission IV of URSI and a Fellow of the Royal Astronomical and Royal Meteorological Societies.

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E. D. Reed (A'48-M'55) was born on October 12, 1919, in Austria. He received the B.Sc. degree in electrical engineering from



E. D. Reed

the University of London, Eng., and the M.S. and Ph.D. degrees from Columbia University, New

York, N. Y. He joined Bell Telephone Laboratories, Murray Hill, N. J., in 1947, where he has been engaged in the development of microwave and millimeter wave tubes

and more recently in parametric amplifiers. He is presently associated with exploratory development of maser amplifiers and microwave ferrite devices.

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Frank Rosenblatt was born in New Rochelle, N. Y., on July 11, 1928. He received the B.A. degree in 1950 and the Ph.D. degree in experimental psychopathology in 1956, both from Cornell University, Ithaca, N. Y.

Since 1955 he has

been on the staff of

the Cornell Aero-

nautical Laboratory,

in Buffalo, N. Y.,

originally as a re-

search psychologist,

and currently as head

of the Cognitive Sys-

tems Section, which

was created in 1959,

He is also a lecturer

in the Department of



F. ROSENBLATT

Psychology and Director of the Cognitive Systems Research Program at Cornell University. During the last five years he has been engaged primarily in the study of mathematical brain models. and the design of tactical control systems.

Dr. Rosenblatt is a member of the American Psychological Association and Sigma Xi.

•••

Robert A. Schmeltzer (S'55-M'56) was born on May 20, 1933, in Brooklyn, N. Y. He received the B.S. degree in electrical en-

gineering from the Cooper Union, New York, N. Y., in 1955, and the M.S. degree in electrical engineering from Stevens Institute of Technology, Hoboken, N. J., in

1958 he was engaged in application engineering, first with the Electron Tube Divi-

R. A. SCHMELTZER

sion, and then the Semiconductor and Materials Division of the Radio Corporation of America, Somerville, N. J. in 1957 he served six months in the U.S. Army doing research and development work at the Signal Corps Engineering Laboratories, Fort Monmouth, N. J. He presently is on leave of absence from RCA as the recipient of the David Sarnoff Fellowship for the second consecutive year, studying full time for the Doctor of Engineering Science degree at New York University, New York, N. Y

Mr. Schmeltzer is a member of Eta Kappa Nu.

Bertram Schwartz (M'56) was born in Brooklyn, N. Y. on November 1, 1924. He received the B.S. degree in chemistry from

New York Univer-sity, New York, in 1949, and has studied physical chemistry in the Graduate School of Columbia University, New York, N. Y. After teaching

chemistry for a short time in a New York city high school, he was employed by the Central Research

Laboratories of Interchemical Corporation, New York, N. Y. to work on problems

World Radio History

B. SCHWARTZ

associated with butyl rubber compounds, In 1952 he joined Sylvania Electric Products, Tewanda, Pa., to work on the preparation of ultra-high-purity silicon. Since 1954 he has been with the Semiconductor Division of Hughes Products, Newport Beach Calif., employed in the areas of surface characterization and chemical etching.

Mr. Schwartz is a member of RESA, the Electrochemical Society, and the American Chemical Society.

L. P. Shkarofsky (M'58) was born in Montreal, Canada, on July 4, 1931. He received the B.S. degree with first class honors



in physics and mathematics from McGill University, Montreal, in 1952. In 1953. he obtained the M.S. degree, conducting his research at the Eaton Electronics Research Laboratory, McGill University, in the fields of microwave optics and antennas. He then joined the microwave

I. P. Shkarofsky

tube and noise group at the Eaton Electronics Research Laboratory, and received the Ph.D. degree in 1957 with a dissertation on modulated electron beams in space-chargewave tubes and klystrons.

During summers, he has worked for the Defence Research Board and for Canadian Aviation Electronics Ltd., in electronics and radar research. After graduation, he joined the Research Laboratories of RCA-Victor Company, Ltd., Electromagnetic Division, where he is presently engaged in research on plasma and microwave physics.

Dr. Shkarofsky is an associate member of the Canadian Association of Physicists.

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I. Torkel Wallmark (A'48-M'55-SM'59) was born in Stockholm, Sweden, on June 4, 1919. He received the degrees of Civilingen-



J. T. WALLMARK

ior in electrical engineering in 1944, Teknologie Licentiat in 1947, and Teknologie Doktor in 1953. from the Royal Institute of Technology, Stockholm.

From 1944 to 1945 he was a vacuum tube designer for the A.B. Standard Radiofabrik, From 1945 to 1953 he was with

the Royal Institute of Technology as a research assistant on vacuum tube problems, while spending periods at the RCA Laboratories in Princeton, N. J., at the Elektrovarmeinstitutet and the Tekniska Forskningsradet, both in Stockholm, engaged in work on secondary emission tubes, semiconductors, and research administration, respectively. Since 1953 he has been with RCA Laboratories, Princeton, where he has been engaged in research on magnetrons, color television, and semiconductor devices, and most recently on integrated electronics.

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Contributors_____

Morrel P. Bachynski (S'54–M'56–SM'58) was born on July 19, 1930, in Bienfait, Saskatchewan, Canada. He received the B.E.



BACHYNSKI

degree in engineering physics in 1952, the M.S. degree in 1953, both from the University of Saskatchewan, Saskatoon, and the Ph.D. degree from McGill University.Montreal, in 1955. Until October

1955, as a member of the staff of the Eaton Electronics Research

Laboratory (McGill University), he was engaged in investigations of aberrations in microwave lens systems. Since that time he has been with the RCA-Victor Research Laboratories, Montreal, Canada, concerned primarily with short-wave propagation and plasma physics problems. He is Associate Laboratories Director of the Microwave Laboratory.

Dr. Bachynski is a member of the Canadian Association of Physicists and Commission VI of the Canadian National Committee of URSI.

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Charles E. Cook (S'49-A'51-M'54-SM'56) was born on October 27, 1926, in New York, N. Y. He received the B.S. degree in physics from



C. E. Cook

Harvard University, Cambridge, Mass., in 1949, and the M.E.E. degree from the Polytechnic Institute of Brooklyn, Brooklyn, N. Y., in 1954. From 1945 to

1946 he served in the U. S. Navy as an electronics technician. Thereafter, an electrical engineer,

he specialized in pulse and electronic control circuits. He joined Sperry Gyroscope Co., Great Neck, N. Y., in 1951, assigned to radar research projects as an assistant project engineer. In 1953 he was advanced to project engineer and in 1956 to senior engineer. He was engaged in the basic investigation of pulse compression and coded transmission techniques, and has served as consultant in this field to other divisions of the company. In 1959 he was promoted to research engineer, where he is at present associated, with further development studies involving the use of pulse compression and correlated techniques applied to radar and communication systems.

Mr. Cook is a member of Sigma Xi.

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W. E. Danielson was born in Fort Collins, Colo., on February 10, 1923. He received the B.S. degree in 1949, and the Ph.D. degree in physics in 1952, both at the California Institute of Technology, Pasadena. He joined the Bell Telephone Laboratories, Murray Hill, N. J., in 1952, where he engaged in microwave

work involving noise

studies, electron beam

formation, traveling-

wave and parametric

amplifiers. In 1958

he became part of

the military effort of

the laboratories at

Whippany, N. J.,

associated with re-

search on pulse com-

pression techniques, antennas, solid-state



W, E. DANIELSON

circuitry and military communications. Dr. Danielson is a member of the American Physical Society, Sigma Xi and Tau Beta Pi.

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Anthonet H. de Voogt (M'49) was born in Amsterdam, The Netherlands, on May 1, 1892. A radio amateur since 1908, he ob-



tained the wireless operator's certificate (class 1) in 1914, and was employed in that capacity on ships of the Holland America line. In 1915 he graduated as an electrical engineer and entered The Netherlands PTT-Services, where he worked in the Long Lines Branch (Cables and

A. H. de Voogt

Repeaters). After World War II he was appointed Head of all PTT Radio Services. When the Radio Services were subdivided in 1952, he became chief of the Ionosphere and Radio-Astronomy section. He installed the first radio telescope in The Netherlands at the Kootwijk radio station.

Mr. de Voogt attended the post-war Plenary Assemblies of URSI, IAU, CCIR, UGGI and CSAGI (International Geophysical Year), and initiated the establishment of geophysical stations at Hollandia, Netherlands New Guinea, Paramaribo, Surinam, and NERA (Nedlerhorst den Berg PTT-Receiving station).

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Tudor W. Johnston (M'59) was born on January 17, 1932, in Montreal, Canada. He received the B.S. degree in engineering physics from McGill

University, Montreal,

in 1953, and the

Ph.D. degree from

the University of

Cambridge, England,

in 1958. While at

Trinity College, Cam-

bridge, he investi-

gated the dynamics

of magnetically-fo-

cused electron beams

from magnetically-



T. W. JOHNSTON

shielded electron guns, including nonlaminar effects and ion phenomena.

He has participated in VHF FM relay development at the RCA-Victor Company, Ltd., Montreal, and on his return from England, joined the Microwave Laboratory of the RCA-Victor Research Laboratories in Montreal. He has collaborated in an analysis of the ion diode rocket and general electrical rocket characteristics, but his chief work is in electron beam and electromagnetic wave interaction with plasmas, and plasma physics in general.

Dr. Johnston is a member of Phi Epsilon Alpha.

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Moises Levy was born in Concepcion, Panama, on April 8, 1930. He received the B.S. degree in chemistry and the M.S. de-



gree in chemical engineering, both from the California Institute of Technology, Pasadena. He also did graduate work in physics at the University of California at Los Angeles, on a cooperative program with Hughes Products.

In 1953 he joined the laboratory of

Specialty Resins Co., Lynwood, Calif., where he was involved in the preparation of alkyds and polyesters. In 1956 he became a member of the technical staff of the Semiconductor Division of Hughes Products, Newport Beach, Calif., eugaging in the study of the influence of surface properties on operating device characteristics. In 1958 he accepted a teaching assistantship at the University of California at Los Angeles, where he is at present.

Mr. Levy is a member of the American Physical Society and Tau Beta Pi.

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Hunter L. McDowell (S'48-A'49-M'55) was born in Washington, D. C., in 1927. He received the B.E.E. degree from Cornell University, Ithaca,

N. Y., in 1948.

From 1948 to 1959 he was employed by The Bell Telephone Laboratories, Murray Hill, N. J., working mainly on the development of traveling-wave tube amplifiers, first for long haul radio relay systems and later for experimental millime-

ter wave systems. In August, 1959, he joined the staff of S.F.D. Laboratories in Union, N. J., where he is engaged in the development of crossed-field amplifiers.

H. L. MCDOWELL

$$dv_{\theta} = \frac{e}{m} \frac{E_{\theta} dr}{v_{r}}$$
$$= -\frac{\frac{e}{m} 2RE_{r} \sin^{2} \theta/2}{\left(2 \frac{e}{m} V\right)^{1/2} \cos \alpha_{1} \sin \theta} \quad (3)$$

The new slope of the path with respect to a radius is, from (1) and (3),

$$\tan \alpha_2 = \frac{v_{\theta_2}}{v_r} = \tan \alpha_1 - \frac{RE_r \sin^2 \theta/2}{V \cos^2 \alpha_1 \sin \theta}, \quad (4)$$

or the change in slope is

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$$\tan \alpha_2 - \tan \alpha_1 = -\frac{RE_r}{V} \sec^2 \alpha_1 \frac{\sin^2 \theta/2}{\sin \theta} \cdot (5)$$

If we call the distance from the polar axis $\rho = R \sin \theta$, then

$$\tan \alpha_2 - \tan \alpha_1$$

$$= \left(-\frac{1}{4V}\right) \left(\frac{1}{\sin^2 \theta}\right) (\sec^2 \alpha_1). \quad (6)$$

In the corresponding cylindrical case,

Fig. 1 represents a section through a cylinder slit along a line parallel to its axis which is perpendicular to the paper. The expression obtained from a similar analysis of this cylindrical case is

$$\tan \alpha_2 - \tan \alpha_1$$
$$= \left(-\frac{\rho E_r}{2V}\right) \left(\frac{\theta}{\sin \theta}\right) (\sec^2 \alpha_1). \quad (7)$$

The first factor on the right-hand sides of each of these equations is the usual expression derived from the Davisson-Calhick lens formulas.2 The second factors containing the θ -dependence show the influence of the geometry, which ordinarily is slight, but which may be significant in some cases. For the flat plane lens, θ becomes zero and both factors become 1. These factors are illustrated in Fig. 2.

The third factor, $\sec^2 \alpha_1$, shown in Fig. 3, is the same in all geometries and shows the influence of the electron's entrance angle, It is interesting to examine more closely the meaning of this factor. Suppose we write the general expression

$$\tan \alpha_2 = \tan \alpha_1 + K \sec^2 \alpha_1, \qquad (8)$$

which can represent either (6) or (7). We now determine the angle α_2 by a Taylor expansion of the inverse tangent of the righthand side of (8) about α_1 .

Klemperer, "Electron Optics," Cambridge University Press, Cambridge, Eug., pp. 62–63; 1953.



Fig. 2-Correction factor for change of slope due to lens curvature



Fig. 3-Correction factor for change of slope due to initial angle of path.

$$\alpha_2 = \alpha_1 + K \sec^2 \alpha_1 \frac{d \tan^{-1} (\tan \alpha_1)}{d \tan \alpha_1} + \frac{K^2}{2} \sec^4 \alpha_1 \frac{d^2 \tan^{-1} (\tan \alpha_1)}{d (\tan \alpha_1)^2} + \cdots$$
(9)

which becomes

$$\alpha_2 - \alpha_1 = K - K^2 \tan \alpha_1 + \cdots . \quad (10)$$

Thus for reasonably small values of K (weak lenses), K gives the change in angle at the leas but not the change in slope of the electron trajectory.

Ideal imaging properties are implied in the usual lens formulas of light optics which are commonly applied in electron optics. Both modifying factors contained in (6) and (7) introduce positive spherical aberration, in effect, making the focal lengths functions of θ and α_1 .

In the usual Pierce gun, where electrons converge along radii, the angle α_1 is ordinarily zero, though θ is not. Since we cannot know the precise surface along which the transverse flux may be considered to act, there is some ambiguity in the above development. Of course one expects aberrations in the focusing of wide-angle cones of electrons,3 but the present treatment should clarify the nature of the influences of curvature and initial angle on the behavior of this particular type of lens.

The author is grateful to his colleagues in the General Electric Research Laboratory for discussions of this note.

L. A. HARRIS General Electric Res. Lab. Schenectady, N. Y.

¹ O. Klemperer, op. cit., pp. 12-14.

On the Origin of Negative Feedback*

Concerning the origin of negative feedback and the interesting application of it by Sargent, brought to light by Grote Reber,¹ there is a prior conception which I think merits a nomination for invention of negative feedback.

On January 9, 1923, a patent application (later to become Serial No. 1.723.719) was filed by Stuart Ballantine on a twotube electrical communication circuit which embodies a feedback network which is adjustable in phase so as to provide "an audio frequency bias voltage in phase with, or exactly opposite in phase to, the modulated radio frequency voltage-passing through to the detector." The primary purpose of this circuit was to reduce the distortion produced by the detector in this reflexed two-tube radio receiver but the patent proceeds to state "that the invention is not limited to a radio receiving circuit of the type shown since a correcting bias voltage of substantially the same wave form as an audio frequency wave or the modulated envelope of a radio frequency wave may be applied to an audion of other amplifying circuits for the purpose of reducing or eliminating the distortion normally present in such a circuit."

Granting that this early picture is complicated by the RF-AF ramifications and the terminology, e.g., audio frequency bias, neutralizing correction-1 would still say that negative feedback was here in 1923.

EDMUND OSFERLAND EO Electronics Inc. Mountain Lakes, N. J.

* Received by the IRE, July 17, 1959, ¹G. Reber, "Negative feedback a third of a century ago," PROC, IRE, vol. 47, p. 1275; July, 1959.

electrodes covering the major surfaces. The piezoelectric stress constant for excitation of this thickness-shear mode as function of the orientation angle θ is given by

$$e_{\theta} = e_{11} \cos^2 \theta + e_{14} \cos \theta \sin \theta. \qquad (1)$$

The same thickness-shear mode can be excited by a field parallel to the plate,1 and the corresponding piezoelectric stress constant is then

$$e_{\theta\psi} = (e_{11}\cos\theta\sin\theta + e_{14}\sin^2\theta)\sin\psi, \quad (2)$$

where the azimuth angle ψ is taken from the X axis in the plane of the plate. For maxinum excitation, the field must be parallel to the Z' axis ($\psi = 90^{\circ}$). In order to provide a field parallel to the plate, the electrodes are arranged so that each electrode covers part of both major surfaces leaving a gap g parallel to the major surfaces of the plate.

Figs. 1 and 2 show laboratory models of quartz crystals of the AT cut vibrating at frequencies of 750 kc and 1000 kc, respectively, and using parallel field excitation. The data of the equivalent electric circuit depends on the width of the gap g. For a bevelled quartz crystal free of unwanted modes, as shown in Fig. 1, vibrating at 750



Fig. 1-Laboratory model of 750-kc quartz crystal excited by a parallel field.



Fig. 2- Laboratory model of a 1000-ke quartz crystal excited by a parallel field.

¹ R. Bechmann, "Über Dickenschwingungen piezoelektrischer Kristallplatten," Arch., elek. Über-tragung, vol. 6, pp. 301-368, September, 1952; Nach-trag, vol. 7, pp. 351-356, July, 1953. R. Bechmann, "Filterquarze im Bereich 7 bis 30 MHz." Arch. elek. Übertragung, vol. 13, pp. 90-93; February, 1959.



. 3---Motional inductance Li of a 750-kc AT quartz oscillator excited by a parallel field as function of Fig. 3the gap g.







Fig. 5—Capacitance ratio $r = C_0/C_1$ of a 750-kc quartz crystal excited by a parallel field as function of the gap g.

kc and having a diameter of 0.825 inch, Fig. 3 gives the motional inductance, L1, Fig. 4 gives the parallel capacitance C_0 , and Fig. 5 gives the capacitance ratio $r = C_0/C_1$ measured as function of the gap g. It has been found that an oscillator excited by a parallel field has a higher value for Q than an oscillator excited by a perpendicular field, provided that these crystals are mounted in a vacuum. The reason for the higher Q is a lower dielectric loss when using parallel field excitation instead of perpendicular field excitation. Because of the very high inductance L_1 and the high values for Q, quartz oscillators excited by a parallel field are perticularly suitable for application to high-precision frequency control. Using wider gaps, the center of the plate where the maximum mechanical stress occurs is not plated and therefore the aging caused by electrodes is reduced.

Results of more detailed investigations will follow in due course.

> R BECHMANN U. S. Army Res. and Dev. Lab. Fort Monmouth, N. J.

Spherical Aberration Due to Initial Path Angle and Lens Curvature in **Aperture Electron Lenses***

The use of large area convergences in beam-type microwave tubes involves fairly large angles and curvatures in the electron guns. It is of interest to inquire into the effect of the spherical or cylindrical geometry on the strength of the anode lens, and into the proper application of the lens formula when an electron passes through the lens at a large angle to the optical axis.

Consider the spherical geometry typical of a Pierce gun as shown in Fig. 1. We make the usual thin-lens approximations by assuming that the radial electric force lines, which would normally fall on the portion of the sphere removed for the aperture, are undisturbed except for the sudden θ deflection in the spherical surface, as shown. The θ -momentum imparted to the electrons as they cross this surface leads to an expression for the lens action.1



Fig. 1-Geometry of a spherical aperture lens.

We investigate the motion of an electron moving in a constant ϕ plane but having both r and θ components of velocity as it enters the lens. If V is the lens voltage these components of velocity are

$$v_r = \left(2 \, \frac{e}{m} \, V\right)^{1/2} \, \cos \alpha_1$$

and

$$v_{\theta_1} = \left(2 \, \frac{e}{m} \, V\right)^{1/2} \sin \alpha_1. \tag{1}$$

The θ -directed force in the lens is found from the deflected radial electric flux, assuming that the field inside R is zero.

$$4\pi R^2 \epsilon E_r \sin^2 \theta / 2 = -\epsilon E_\theta 2\pi R \sin \theta dr$$

or

$$E_{\theta}dr = -2RE_r \frac{\sin^2 \theta/2}{\sin \theta}$$
 (2)

* Received by the IRE, September 18, 1959. 1 L. A. Harris, "The electron optical action of an annular aperture lens," PROC. IRE, vol. 46, pp. 1655-1656; September, 1958.

Correspondence

No. 2, respectively, when focused at infinity.

Note that both equations are subject to the constraint $D_1 + D_2 = C_{\min}$. Note also that the value of K_T depends only on the aperture shape and distribution function. Typical values of K_T are as follows.⁴

Uniform square aperture	$K_T = 1$	S.L. =13.2 db
Uniform circular aperture	$K_T = \frac{\pi}{4\sqrt{2}}$	S.L. =17.6 db
(1 −ρ ²) tapered ⁵ circular aperture	$K_T = \frac{\pi}{8}$	S.L. =24.6 db

An alternative form of the optimization relation is

$$D_1 D_2 = \frac{\lambda R}{\sqrt{K_{T_1} K_{T_2}}} \tag{3}$$

which, when combined with $D_1 + D_2 = C_{\min}$, yields

$$\begin{bmatrix}
 D_1 \\
 D_2
 \end{bmatrix} = \frac{1}{2} \left[C_{\min} \pm \sqrt{C_{\min}^2 - \frac{4\lambda R}{\sqrt{K_{T_1} K_{T_2}}}} \right]$$
(6)

From (4) it is obvious that D has an imaginary component for

$$C_{\min}^2 < \frac{4\lambda R}{\sqrt{K_T, K_{T_*}}},$$

since both C_{\min}^2 and

$$\frac{4\lambda R}{\sqrt{K_{T_1}K_{T_2}}}$$

are always real and positive. The interpretation of this imaginary component of apertures embraces what is known in antenna theory as "supergain." That is, for

$$C_{\min} < C_0 = \left[\frac{4\lambda R}{\sqrt{K_{T_1}K_{T_2}}}\right]^{1/2},$$

the antennas must be conjugately supergained in order to be optimally coupled. Since antenna engineers know that supergain is practical only in theory due to high thermodynamic losses or, alternatively, astronomical values of Q, it may be concluded that values of C_{\min} less than C_0 cannot be used for efficient power transmission. Note, however, that most "intelligence" transfer systems operate satisfactorily with values of C_{\min} much less than C_0 without resorting to supergain. Operation in the latter mode is quite adequately described7

by the so-called "one-way transmission equation" or "one-way radar equation." The final design equations are then

$$C_{\min} = \left[\frac{4\lambda R}{\sqrt{K_{T_1}K_{T_2}}}\right]^{1/2}$$
(5)

and as a consequence,

$$D_1 \equiv D_2 = \frac{1}{2} C_{\min}$$
 $K_{T_1} \equiv K_{T_2}$. (6)

Unfortunately, since the above system couples only the main beams down to about their respective half-power points, the transfer efficiency is fairly low. The value depends somewhat on the aperture distribution,8 because this in turn controls the relative energies in the main beam and sidelobe complex of an antenna. Typical efficiencies are ideally about 45 per cent for the uniform circular aperture and about 72 per cent for the $(1-\rho^2)$ tapered aperture.

$$\frac{1}{2} = \frac{1}{2} \left[C_{\min} \pm \sqrt{C_{\min}^2 - \frac{4\lambda R}{\sqrt{K_{T_1} K_{T_2}}}} \right]$$
(4)

FOCUSED ANTENNAS

In a paper referred to in the Introduction,3 the author established the concept of focused antennas as a means of exceeding the performance of conventional antennas when operated under near-zone conditions. Such operation requires a large ratio of aperture diameters but can provide coupling efficiencies substantially greater than the conventional antenna system. The reason is that each antenna will essentially "see" more than the main beam of the other antenna without experiencing the degradation in performance normally caused by Fresnel interference.

In this system, the smaller antenna may be conventional and should be of a size

$$D_2 = \sqrt{\frac{\lambda \overline{R}}{K_{T_2}}} \quad . \tag{7}$$

The larger antenna, D_1 , should be as large as practical and focused on the center of phase of D_2 . The coupling efficiency is then approximately

$$\eta = \eta_1 \eta_2,$$

where η_1 is the ratio of the energy in the portion of the Fraunhofer pattern of D_1 seen by D_2 to the total spacial energy and η_2 is the corresponding quantity for D_2 . An ideal efficiency of 96 per cent can be obtained for $(1-\rho^2)$ tapered antennas when $D_1 \approx 2\sqrt{\lambda R}$ and $D_2 \approx 1.6 \sqrt{\lambda R}$.

⁸ R. C. Hansen, "Low noise antennas," Micro-wave J., vol. 2, pp. 19-24; June 1959.

An optimum relation, in terms of diminishing returns, would be

$$D_2 = \sqrt{\frac{\lambda R}{K_{T_2}}}$$
$$D_1 = A \sqrt{\lambda R K_{T_2}}$$
$$K_{T_1} = K_{T_2}$$

where $A = D_2/\lambda$ {Null to Null beamwidth of D_2 in radians}

Distribution	А	
Uniform Square	2.00	
Uniform Circular	2.44	
$(1 - \rho^2)$ Tapered Circular	3.26	

Note that the over-all circuit is a network problem⁹ and that the foregoing applies only to the mutual coupling elements of this network.

Note also that power levels greater than the capacity of a single transmission circuit should be handled by multiple circuits operated in an incoherent (e.g., different frequency) manner.

Finally, it should be mentioned that the total amount of waveguide in an identical pair of two-dimensional slot arrays, which are optimally coupled, is approximately four times the amount needed to run a direct transmission line between the two sites. Consequently, the reason for transferring power via radio waves must be more esthetic than simply the desire to transmit high frequency energy from A to B in the most economical manner.

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9 S. Tetelbaum, "On the problem of efficient longdistance wireless power transmission," (U.S.S.R.), vol. 9, p. 505-514; June, 1945.

Improved High-Precision Quartz **Oscillators Using Parallel** Field Excitation*

Piezoelectric oscillators which have the form of plates and bars and vibrate in various modes of motion, e.g., thickness modes, contour modes or extensional modes, generally can be excited by a field perpendicular to, or by a field parallel to, the major surfaces. Of particular interest is the behavior of quartz plates of the orientation $(YXl)\theta$ vibrating in the thickness-shear mode xy', e.g., the AT or BT cut. The usual excitation of these cuts is achieved by a field perpendicular to the plate using two

* Received by the IRE, September 18, 1959,

⁴ R. W. Bickmore and R. C. Hansen, "Antenna power densities in the Fresnel region," Proc. IRE, to be published.
⁵ S. Silver, "Antenna Theory and Design," M.I.T. Rad. Lab. Ser., McGraw-Hill Book Co., Inc., New York, N. Y. no, 12, p. 194–195.
⁶ R. W. Bickmore, "A note on the effective aperture of electrically scanned arrays," IRE TRANS, ON ANTENNAS AND PROFAGATION, vol. AP-6, pp. 194–196; April, 1958.
⁷ S. Silver, op. cit., p. 4 S. Silver, op. cit., p. 4.

Neither formula, as given above, can be expected to give a correct answer. Presumably, this is due to the fact that flux leakage effects and current sheet corrections are not considered in these equations. Various correction factors are usually applied to (1), and an attempt was made to evaluate the validity of these correction terms in the previously mentioned work at the National Bureau of Standards. However, one result that came out of this study, which has not been sufficiently emphasized, was that the errors in these measurements appeared to be largely independent of the magnitude of the permeability. This is apparently true even for a permeability of the order of unity which would be the case for the polystyrene core. It thus follows that errors in permeability measurements should subtract out if $L_{\rm m}$ and $L_{\rm r}$ are both measured and the results are substituted in an equation of the following form:

$$\mu' = \frac{L_m - L_s}{L_a} + 1.$$
(3)

This equation follows readily from (1) if we let ΔL be the error in the measured value of both L_m and L_s . Since ΔL is independent of permeability, we should write for the correct values of L_m and L_s :

$$L_m = \mu' L_a + \Delta L \tag{4}$$

and

$$L_s = L_a + \Delta L. \tag{5}$$

Subtraction of (5) from (4) will lead directly to (3).

In order to obtain a comparison of (1) through (3), a series of measurements were made at 50 kc using a Maxwell-type inductance bridge. A powdered iron core was used having a permeability of 7.05 as determined by a radio-frequency permeameter. The resulting data giving permeability as a function of number of turns used on the coil, are given in Fig. 1.

The increase in accuracy obtained using (3) is readily apparent. It was also observed that a curve similar to that resulting from (3) could be obtained by using (1) alone with corrections which account for finite wire size and departure of the wire from a uniform current sheet. On the other hand, the calculation of these corrections is usually quite tedious. As such, utilization of (3) offers a distinct advantage even though it does make it necessary to wind a separate coil on a polystyrene core. However, this must be done anyway whenever a measurement of magnetic loss is desired.

The above measurements were made using coils wound directly on the samples. Data are also given in Fig. 1 using (3) for the case where the sample has been reduced in size and a 0.05-inch layer of polystyrene lies between the core and the winding. The over-all cross section dimensions of the coil are the same as used in the previous measurements. A slight modification in La must, of course, be made in order to account for the reduced size of the core. As can be seen in the graph, removing the wire somewhat from the surface of the sample improves the accuracy of permeability measurements for a small number of turns. Presumably this is due to the creation of a more uniform flux



Fig. 1—Relationship between permeability, inductance formula, and number of turns on toroidal coil (inside coil dimensions: I,D. =0.77 inch, O,D. = 1.37 inches, height =0.30 inch, No, 36 wire).

inside the sample. The technique of separating the winding from the sample has been used in the past by other workers for obtaining more accurate loss measurements; however, the resulting improvement in real permeability apparently has not been adequately emphasized.

The above results indicate that coils with a small number of turns can be constructed which can be opened for sample insertion and which will take advantage of the accuracy and convenience of inductance measurements obtained in connection with (3). Such coils have been described in the literature.³ Similar work at the National Bureau of Standards has shown that coils of this type are satisfactory for precise measurements of both μ' and magnetic loss in the frequency range up to several mc.

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⁴ I. Epelboin, "A study of metals with Hertzian waves with the aid of demountable winding permeameter," L'Onde Electrique, vol. 28, pp. 322-327; August-September, 1948.

Power Transmission via Radio Waves*

INTRODUCTION

The current revival of interest in the problem of transferring sizable amounts of power via radio link^{1,2} has raised some

* Received hy the IRE, September 14, 1959. ¹ M. T. Willinski, "Beamed electromagnetic power as a propulsion energy source," *J. Amer. Rocket Soc.*, vol. 29, pp. 601-603; August, 1959. ² "Raytheon microwave powered helicopter," *Time*, vol. 73, p. 55; May 25, 1959. questions as to the optimum antenna system which can be designed. In view of the fact that the antenna system represents the essence of radio power transmission, the main purpose of this letter is to indicate a design criterion which optimizes such an antenna system with due regard to the law of diminishing returns. The theory of coupled antennas has been presented in a previous paper³ and consequently only the results, which have been adapted to this problem, are presented here.

CONVENTIONAL APPROACH

Consider two coupled pencil beam antennas, one transmitting and one receiving. Free space propagation will be assumed as well as no depolarization loss. The receiving antenna is assumed to be conventional in that it concentrates the received energy onto a "terminal pair" as opposed to being a large absorbing mass through which the heat exchanger fluid passes.

The geometry is shown in Fig. 1.



In this problem, a constraint will be placed on the antennas: namely, D_1+D_2 = minimum. This is a reasonable engineering as well as economic constraint and, in addition, allows for an optimum solution as opposed to an unrestricted maximum solution which would be devoid of all physical meaning.

The optimization condition may be stated several ways as follows:

$$K_1 K_2 = K_{T_1} K_{T_2}$$
(1)

or

$$R_{T_1}R_{T_2} = R^2 \tag{2}$$

where

$$K_1 = \frac{\lambda R}{D_1^2}$$
$$K_2 = \frac{\lambda R}{D_2^2}$$

$$K_{T_1}$$
 = value of K_1 at $R = R_{T_1}$

 K_{T_2} = value of K_2 at $R = R_{T_2}$

 $\lambda =$ wavelength

D = diameter or length of side of an antenna.

 R_{T_1} and R_{T_2} are the Fresnel-Fraunhofer transition distances of apertures No. 1 and

⁸ R. W. Bickmore, "On focusing electromagnetic radiators," *Can. J. Phys.*, vol. 35, pp. 1292-1298; November, 1957.

If the region of observation is in the vicinity of the apogee or perigee of the orbit and the frequency is sufficiently high, dependent on the electron density level existing at the time of the measurement, then, for a plane earth approximation, the electron content below the satellite may be determined from the expression1

$$N_h = \frac{\pi h f^2}{2.97 \times 10^{-2} H v T},$$

1)

where

 $N_{h} = \text{electron content}$ in a column of height h and 1 square meter in area,

h =satellite height in meters,

f = frequency in cps,

- H = mean value of the earth's magnetic field in the direction of motion of the satellite in amps/meter,
- v = the satellite velocity in meters/sec,
- T = time between adjacent nulls of the Faraday rotation in seconds.

The value of the earth's magnetic field used to compute the electron content is determined on the basis of an assumed Chapman distribution of electrons with height, and from tables of the earth's magnetic field variation with height.2 Thus, a mean value of magnetic field is obtained which is dependent on the satellite height.

Values of electron content obtained during the period March, 1959 to May, 1959 for satellite passes meeting the requirements for the approximations used in deriving (1) are plotted as a function of local time in Fig. 1. Unfortunately, the satellite height for each measurement was not constant and, for the data shown, varied between 1227 and 1350 km. Thus, some scatter of the measured points would be expected because of this height variation, but as the electron density is relatively low at these heights the scatter is probably small compared with the daily variations in the level of electron density shown by conventional ionospheric sounding apparatus. The highest values of electron content occur, as expected, in the early afternoon with a range of 8.5 to 5.5 ×1017 electrons. A mean value of 6.9×1017 electrons is obtained for the period between 1330 and 1630 hours.

Earlier measurements during November, 1958 to January, 1959 for the near midnight period are shown in Fig. 2. The variation of the satellite height is greater in this case but its effect on the electron content will be less than for the davtime measurements as indicated later by the change in electron distribution. A mean value of 1.1×1017 electrons is obtained near midnight.

From the data shown in Fig. 1 and Fig. 2, it can be estimated that variations in electron content of 6:1 can occur between the early afternoon and near midnight periods of the day.

Of considerable importance in estimating the variation of electron density with height above the earth is the electron distribution ratio defined by

critical frequency data3 for the Washington area (approximately at the same latitude as

Number of electron above the F_2 layer maximum density Nf + R =Number of electrons below the F_2 layer maximum density Nfa-

As the number of electrons below the F_2 maximum can be obtained from ionospheric soundings from the earth, the electron distribution ratio may be found from the relationship

$$R = \frac{N_h - N_{f_2^-}}{N_{f_2^-}}$$

where N_h is determined from (1) for a suitable satellite orbit.

Figs. 3 and 4 show the electron distribution ratio obtained for the midnight and noon periods during November, 1958 to January, 1959. The number of electrons below the F_2 maximum was calculated from



Fig. 1 -Electron content during the period from March, 1959 to May, 1959,



Fig. 2—Electron content near midnight during the period from November, 1958 to January, 1959.



Fig. 3 —Electron distribution ratio near midnight dur-ing the period from November, 1958 to January, 1959.



Fig. 4—Electron distribution ratio near midday dur-ing the period November, 1958 to January, 1959.

Columbus) and the assumption of typical distribution of electrons in the F_2 , F_1 , E, and E_s regions.⁴ The mean value of the electron distribution ratio is found to be 3.1 during the noon period and 0.96 during the midnight period. Therefore, it is concluded that the distribution of electrons with height above the F_2 layer maximum varies considerably during the midnight to noon period.

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^a "Detailed Values of Ionospheric Characteristics and *F*-plots for Washington," Central Radio Propaga-tion Lab., Natl. Bur, of Standards, Boulder, Colo. ⁴ G. H. Millman, "An Analysis of Tropospheric, Ionospheric and Extraterrestrial Effects on V.H.F. and U.H.F. Propagation," General Electric Co., Syracuse, N. Y., T.I.S. Rept. No. R56EMH31; 1956.

A Technique for Reducing Errors in Permeability Measurements with Coils*

A recent note¹ gave a summary of the results of an investigation carried on at the National Bureau of Standards by Kostyshyn and Haas² for the purpose of studying the sources of error that are associated with the measurement of initial permeability using coils wound on toroidal samples of rectangular cross section. It was pointed out that various laboratories are, in general, using one of the following formulas for evaluating the permeability of a magnetic material.

$$\mu' = \frac{L_m}{L_a} \tag{1}$$

01

$$\mu' = \frac{L_m}{L_s} \tag{2}$$

where

- μ' = real part of the relative initial complex permeability,
- L_m = measured inductance of the coil wound on the magnetic sample,
- L_a = the calculated inductance of an equivalent air core coil assuming a thin uniform current sheet,
- L_s = measured value of inductance of an identical coil wound on a polystyrene core.

* Received by the IRE, September 14, 1959.
¹ R. D. Harrington and R. C. Powell, "High-frequency magnetic permeability measurements using toroidal coils," PRoc. IRE, vol. 46, p. 784; April, 1958.
² B. Kostyshyn and P. H. Haas, "Discussion of current-sheet approximations in reference to high-frequency magnetic measurements," J. Res. Natl. Bur. Standards, vol. 52, pp. 279–287; June, 1954.

¹ T. G. Hame, and W. D. Stuart, "The Faraday Rotation of Radio Transmissions from Earth Satel-lites and the Electron Density in the Ionosphere," Antenna Lab., The Ohio State Univ, Res. Found, Columbus, Rept. 889-5; August 1959. ² E. H. Vestine, et al., "The Geomagnetic Field, Its Description and Analysis," Carnegie Institution of Washington, Washington, D. C., publication 580; 1947.

^{1947.}

known that for the above simple type of nonlinearity there always appear stable and unstable singularities along the nonlinear characteristic in consecutive order, and here we may rule out p_1 and p_{-1} as unstable. The proof of this statement follows the usual perturbation approach and is readily obtained. From an analogous argumentation, we also can rule out p_1 , p_{-1} as possible values in our step-functions, since otherwise disturbances would grow in the time intervals corresponding to these values. Making use of the special properties of our step-functions, this statement may also be verified analytically by methods analogous to some to be presented later.

Thus we can expect a stable solution of our system, as shown in Fig. 3.



Although this type of function is an arbitrary one to a large extent, it demonstrates some basic properties of physical systems.

- 1) The spectral lines of the corresponding energy distribution are spaced $2\pi/\tau$ radians, and it is easily seen that they coincide with the mode frequencies defined in (6).
- 2) The fact that $p_0=0$ constitutes a stable solution leads to pulse character of the waveform, i.e., zero amplitude may be maintained during portions of its duration. This is the significant difference from a system with small signal loop-gain, which will lead to square-wave type waveforms where zero amplitude cannot be maintained during any finite time interval.

Stability of our general solution, P(t), is readily shown through the following argumentation: Assume a small perturbation function δp such that

$$e_1(l) = P(l) + \delta p. \tag{9}$$

Transmission of this combination through the nonlinear fourpole renders

$$e_{2}(l) = \mu [P(l) + \delta \rho] + \rho [P^{3}(l) + 3P^{2}(l)\delta \rho] + \eta [P^{5}(l) + 5P^{4}(l)\delta \rho], \qquad (10)$$

where small quantities are neglected. It is easily realized from inspection of Fig. 3 that

$$P^{2}(l) = \left\langle \frac{p_{2}^{2}}{0} \right\rangle = \text{const.}, \qquad (11)$$

and thus we have⁵ a disturbance term:

⁶ Since $0 \le \mu < 1$, for all cases of interest here, no unstable disturbance δe_2 can arise for $p^2(t) = 0$ in (11).

$\delta e_2 = (\mu + 3\rho p_2^2 + 5\eta p_2^4) \delta \rho.$ (12)

The coefficient of δp in (12) is independent of time, due to the special properties of the otherwise largely arbitrary function P(t) recognized above.⁶ A necessary condition for stability of P(t) is that the absolute value of the above coefficient is less than unity, which condition may be expressed by the inequality.

$$\rho p_2^2 - 2(1 - \mu) > 0, \tag{13}$$

where

$$p_{2}^{2} = -\frac{1}{2\eta} \left[\rho + \sqrt{\rho^{2} + 4\eta(1-\mu)} \right].$$
(14)

Eqs. (13) and (14) specify the separatrix

$$p^{2} + 4\eta(1 - \mu) = 0, \qquad (15)$$

which is immediately recognized as the condition that (6) has five distinct, real roots. Thus, P(t) is always stable if such roots may be found for the case of the fifth-order polynomial which represents a good approximation of physical cases. As shown in Fig. 3, the idealized system will indefinitely preserve a waveform, which has been forced upon it by initial conditions, for example, by an external signal $e_i(t)$. Corresponding to the infinite number of degrees of freedom of our system, this waveform is seen to be of rather arbitrary character. In the following, we shall attempt to discuss some properties of physical systems, by combining the results just obtained with previous ones and by using empirical reasoning.

DISCUSSION OF STABILITY AND CONCLUSIONS

The most important factor which enters the considerations for physical systems is bandwidth. Thus, we will always have a finite set of mode frequencies, and it seems plausible to assume that all modes have to contribute to the steady state. This is due to harmonic and combination frequency generation in the nonlinear expander. Also, the amount of dispersion in the loop, which, for convenience, may be attributed to the transfer phase of the feedback network, will be an important influence on stability and the shape of the waveform. Stability tests for physical systems thus have to take into account phase shift, the discussion of which would exceed the present space.

For the case of our special interest, the recirculative RF pulse generator, we have to accept the hypothesis that a "bandpass transformation" is permissible, which makes our previous results applicable to the waveform envelope. Again our set of mode frequencies will define the spectral energy distribution, all modes contributing to the steady state. Pulse length should thus correspond to approximately twice the reciprocal bandwidth, unless jumps in carrier frequency occur. (This phenomenon should only be possible for multiple hump characteristics, and is rather unlikely to appear for a loop transmission peaked at the center and decreasing smoothly toward the edges of the band.)

As shown by Cutler's work,1 a system

* This is directly true for square-wave type func-tions only, which do not assume zero value for any finite time. However, one can convince oneself that stahility also holds if the waveform does assume zero value, as in pulse-trains.

with restricted bandwidth and a frequency characteristic of the general nature of the Gaussian type generates pulses of approximate length 2/Af. Our idealized system was capable of sustaining any group of rectangular pulses, as long as they were periodic in τ , and we could venture to assume that this is also true for restricted bandwidth, for pulses of correspondingly defined length and "rounded off" shape. Thus, such a system could serve as a regenerative pulse memory, using binary pulse/no pulse or phase script (where a 180° carrier phase shift distinguishes between binary states).7

Experimental evidence obtained in the laboratory demonstrated the possibility of recirculation of several pulses within each interval⁸ τ . This may be accomplished by periodic gain variations within the feedback loop.9 For example, if a sinusoidal gain variation with period τ is employed, one obtains single pulse recirculation. For gain variations with a period τ/n , one could circulate n pulses, which constitutes the possibility of introducing a clock-frequency for memory applications. Also, certain phase correction methods should represent another possibility of forcing stability of special wavetrains in the system, for phase-script applications.

Although most of the conclusions which have been presented were obtained following a somewhat intuitive approach, there is sufficient experimental evidence that they definitely should be valid. Specific evaluation to obtain rigorous conclusions would lead to the analytical path pointed out by Cutler, which is difficult at best. Still we could demonstrate the largely arbitrary character of steady-state waveforms to be expected, and outline a procedure for appropriate stabilization.

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⁷ M. P. Forrer, L. Fein, and V. Met, "Microwave Techniques for Computer Applications," Quarterly and Final Report on ONR Contr. NONR-2127(00); December, 1956 and December, 1958.
 ⁸ In Cutler's experiments, an automatic gain-con-trol was employed to obtain single-pulse recirculation.
 ⁹ S. Frankel, private communication.

The Electron Content and Distribution in the Ionosphere*

Recordings of the Faraday rotation rate on the radio transmissions of the satellite 19585 2 at frequencies of 20.005 and 40.01 me have been analyzed to determine the electron content and distribution in the ionosphere. During the period of observation from November, 1958 to May, 1959, for passes in a north to south direction near Columbus, Ohio, the satellite height has varied within the range of 1227 to 1545 km.

^{*} Received by the IRE, August 17, 1959. This work was partially supported by the USAF through the Wave and Propagation Group, Wright Air Dev. Center, Wright-Patterson Air Force Base, Dayton, Oitio.



Fig. 3—Gain with pump on, and insertion loss with pump off plotted against input signal level. Con-ditions as in Tables II and III; pump power into pump cavity 107 nw.



Fig. 4—Gain for signal and idler with pump on, signal loss with pump off, and electronic gain plotted against frequency. Conditions as in Tables II and III; pump power into pump cavity 107 mw,

that the bandwidth is not limited by the gain mechanism in the pump cavity, but simply by the input and output couplers. In the present case it is the Q of the coupling cavities rather than the coupling mechanism to the cyclotron wave that determines the bandwidth. The loaded Q of the cavity with the beam on is about 100 for a matched input. This yields a calculated bandwidth of 41 nnc. The larger measured bandwidth is probably due to accidental detuning of the output cavity as mentioned above. A relative shift of 37 mc between input and output cavities gives a calculated bandwidth of 67 mc and a 1.5-db transmission loss.

Noise has so far been measured at the band center only, within a bandwidth of 2 mc. The best result obtained has been an effective input noise temperature of 225°K, corresponding to a noise figure of 2.5 db for double band working.

The authors thank Dr. R. Kompfner for suggesting this work and A. J. Rustako, Jr. for constructing much of the apparatus.

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On the Regenerative Pulse Generator*

Experimental as well as theoretical work on the millimicrosecond pulse generator has been reported by Cutler.¹ Also, Edson² has pointed out the possibility of interpreting Cutler's circuit as a multimode oscillator. In the following, a more detailed analysis based on this point of view will be presented. For this, we shall assume a circuit as illustrated by the block diagram of Fig. 1. The proper choice of external signal, external gain variation, and nonlinear characteristic, as well as transmission of the feedback network, will assure a maximum variety of possible steadystate waveforms.



Fig. 1--Block diagram of a regenerative pulse generator, or multimode oscillator.

Upon closer investigation of the general problem, one realizes that the strictly academic assumption of an ideal, flat system with an infinite number of modes will permit us to obtain some interesting results with relative ease. Thus we discard the feedback network, (Fig. 1), which provides frequency selectivity in the loop, and accordingly have

$$e_3(l) = e_2(l).$$
 (1)

With the feedback equation

$$e_1(t) = e_2(t - \tau), \qquad (2$$

and with the nonlinear characteristic

$$e_2(t) = f[e_1(t)],$$
 (3)

we obtain a nonlinear difference equation to describe our system. (For our analysis we disregard forced oscillations, and assume $e_i(t) = 0,)$

Although there will be no small signal transmission in an ideal expander, the mode frequencies obtained from a linearized analysis³ as

$$\omega = \frac{2n\pi}{\tau} \tag{4}$$

* Received by the IRE, July 13, 1959.
¹ C. C. Cutler, "The regenerative pulse generator,"
PROC. IRE, vol. 43, pp. 140–148; February, 1955.
² W. A. Edson, "Nonlinear Effects in Broadband Delay Type Feedback Systems," *Proc. Symp. on Nonlinear Circuit Analysis*, Polytech. Inst. of Brooklyn, N. Y., pp. 41–53; April 25–27, 1956.
³ V. Met, "On multimode oscillators with constant time delay," PROC. IRE, vol. 45, pp. 1119–1128; August, 1957.

will also have significant meaning in the present case.4

The difference equation obtained from combination of (2) and (3) has periodic solution functions P(t) with a period τ , and thus we can prescribe:

$$e_1(t) = P(t) = P(t + m\tau).$$
 (5)
(m = 1, 2, 3 · · ·).

Specifically, by substituting (3) into (2), and assuming a solution (5) as defined above, we readily obtain

$$P(t) = f[P(t - \tau)] = f[P(t)].$$
(6)

Eq. (6) has to hold at all times, and for $f(e_1)$ being a polynomial in e_1 , we obtain an algebraic equation in P(t), which may have real solutions:

$$P_1(t) = p_1$$

$$P_2(t) = p_2$$

$$\cdots$$

$$P_n(t) = p_n.$$
(7)

The p_{ν} are constants, or functions which are trivially periodic in τ , and it is easily seen that a step-function which alternates between the various values p_{ν} and which is periodic in τ , is the general solution of our system. A single sinusoid, or a finite combination of sinusoids, cannot be a solution of the system of infinite bandwidth, assumed for our considerations, due to generation of harmonic and combination frequencies in the nonlinear fourpole.

The solutions of (6) are conveniently visualized by a graphical representation, as shown in Fig. 2.





For our investigations of expander action, the nonlinearity is assumed to be represented by a fifth-order polynomial containing odd powers only, as specified by

$$e_{2}(l) = f[e_{1}(l)]$$

= $\mu e_{1}(l) + \rho e_{1}^{3}(l) + \eta e_{1}^{5}(l).$ (8)

We may find five real solutions, $p_{\rm P}$, corresponding to the intersection with the feedback line, as indicated in Fig. 2. It is well

⁴ The phenomenon mentioned by Cutler⁴ that a 180° phase shift in the loop shifts the lines observed on a spectrum analyzer, with the circuit in operation, by halt their spacing in frequency, is readily attributed to basic properties of multimode oscillators. Such phase reversal changes the resonance length of the loop by one-half a wavelength, and a corresponding shift of all mode-frequencies by approximately half the mode-spacing at center band results.



Fig. 1-Microwave Adler tube for 4000 mc.

TABLE 1 TUBE DIMENSIONS 1) Input and output cavities 0.53 inch Plate length Plate width Plate spacing Cavity diameter Cavity length 0.035 Inch 0.043 inch 0.8 inch 0.76 inch 2) Pump cavity Plate length 0.125 inch 0.046 inch 0.40 inch 0.38 inch Plate spacing (along a diameter) avity diameter Cavity length 0.015 inch 3) Beam diameter

TABLE 11 Electrical Characteristics

13	Input and output cavities		
• /	Resonant frequency	4140	me
	Unloaded O	1820	
	Unloaded shunt impedance	87,500	ohms
	Loaded O (cold)	180	
	Input VSWR (cold)	9.2	
2)	Pump		
	Resonant frequency	8300	mc

TABLE III Typical Operating Characteristics

1480 gauss
17.5 volts
70 microamps
107 mw
19.5 db
4107 to 4174 mc
= 67 mc
0.2 db
0.56 db each
80 db

tudinal magnetic field. The pump cavity has four poles and is operated in the π mode. producing a quadrupolar field. It is resonant at the pump frequency (8280 mc) which is approximately twice the signal frequency. The cavities are fed by coaxial lines and loops. The whole tube is immersed in a uniform longitudinal magnetic field of 1480 gauss. A standard 3-anode low-noise traveling-wave tube gun is used. In most of the tests the cathode diameter was 0.026 inch, but the beam diameter was limited to approximately 0.015 inch by apertures in two of the gun anodes. Similar results have been obtained with a 0.015-inch diameter cathode and no limiting apertures. Typical values of beam voltage and current are 17 volts and 70 microamps. The collector is specially designed to prevent secondary electrons returning down the tube.

The operation of the tube is very similar to that of the low frequency tube.¹ With correct adjustment, the input and output cavities act as nearly perfect couplers to a fast cyclotron wave of infinite phase veloc-



Fig. 2—Gain plotted against the square root of pump power. Conditions as in Tables II and III; signal input level -90 dbm.

ity on the beam. In this wave each electron executes an orbiting motion at the cyclotron frequency with all orbits having the same radius and phase. The input cavity feeds all the signal power (circuit losses apart) onto the beam and strips all the noise signal from the beam. The output cavity strips the amplified signal from the beam and feeds it to the output line. In the absence of pump power, signal power is transferred from the input to the output line with some small amount of loss in the cavities. With the pump power on, the signal wave is amplified in the pump cavity, and at the same time an idler wave is generated, which at high gain is of the same amplitude as the signal wave. The extra power in the amplified waves comes, of course, from the pump, and not from the dc beam. The frequency of the idler is the difference between the pump and signal frequencies. If both signal and idler frequencies are within the operating band of the output coupler, both will be stripped off and fed to the output line.

Some of the measured electrical characteristics of the tube are given in Table II and some typical operating conditions in Table III. Fig. 2 shows measured gain plotted against the square root of pump power. In this measurement, signal and idler frequencies were practically equal and both were accepted by the detecting receiver which had a bandwidth of 2 mc. According to theory^{1.3} the maximum db gain in true degenerate operation is strictly proportional to the square root of pump power at all values of pump power. Because of cavity and other losses, the gain at zero pump power will in practice be slightly negative. When

*W. H. Louisell, private communication.

March

the pump is not phase locked to the signal, the db gain is also proportional to the square root of pump power at high gain, but there is a 3-db penalty (taking signal and idler together) and a 6-db penalty for signal alone. In the present experiment, therefore, the gain points at high gain should lie on a straight line, but the point for zero pump power should lie on the line which would be obtained for degenerate operation, 3 db above this. Fig. 2 shows that below saturation the behavior is roughly as predicted. Of the loss at zero pump power about 1.4 db is due to circuit loss. The rest is probably due to accidental detuning of the output cavity.

Saturation occurs above 20 db net gain, due to beam current interception. Up to the point indicated on the curve, the measured interception is zero, but above this point it suddenly starts to rise. A possible explanation of this follows from the ideas of Lea-Wilson⁴ concerning noise in Adler tubes. He points out that the random noise orbits of the electrons leaving the cathode are not removed by the input coupler; only an imperceptible rearrangement occurs, whereby the noise which can be coupled to an external circuit is reduced to zero. These noise orbits can be amplified in the same way as the signal; if they get too large, electrons will strike the cavity posts causing interception and loss of gain.

If the cathode temperature T is 1000° K the average transverse energy of emitted electrons is kT (where k is Boltzmann's constant) or 0.086 volt. Four per cent of the electrons have a transverse energy exceeding 2kT or 0.172 volt. In the present case, this gives an orbit radius before amplification of 2.7×10^{-4} inches for kT electrons and 3.8×10^{-4} inches for 2kT electrons. From the electronic gain at the point of incipient structure interception one can readily compute1 the maximum orbit radius after amplification. It is 5.4×10^{-3} inches for kT electrons and 7.6×10^{-3} inches for 2kT electrons. The nominal clearance between the beam and the cavity posts is 14×10^{-3} inches. It would appear, therefore, that expansion of the beam due to amplification of noise orbits is of the right order to explain current interception and gain saturation.

Gain saturation due to a CW signal is shown in Fig. 3. Saturation starts at a signal input of -20 dbm with a total gain of 22.5 db, corresponding to a signal gain of 19.5 db and an output signal power of 8.9×10^{-4} watts. The calculated² orbit radius corresponding to this power is 3.2×10^{-3} inches. The maximum orbit radius due to signal and idler together is double this figure, or 6.4×10^{-3} inches. This is approximately the same size as the noise orbits which produced saturation in the previous experiment. It seems probable, therefore, that signal saturation is also a result of beam expansion due to amplification of the orbits.

The bandwidth of the tube is shown in Fig. 4. The measurement receiver had a bandwidth of 2 mc. The bandwidth to points 3 db below the gain at band center is 67 mc. It is important to note that the electronic gain is constant with frequency. This means

[°]C. P. Lea-Wilson, "Some possible causes of noise in Adler tubes," PROC. IRE, vol. 48, pp. 255-256; February, 1960.

although wet oxygen shows large discrepancies at around 400 mc. In fact, wet oxygen appears to go through a minimum in this frequency range, a fact which is not presently understood.

If one considers the surface resistance R_{s1} to be the same as the dynamic lowfrequency barrier resistance, a serious objection to the above model is raised when one calculates the values of R_{s1} from the data (see Table II), for it is seen that R_{s1} is about two to three orders of magnitude lower than the low frequency dynamic resistance (approximately 10 megohms at -1volt bias). Furthermore, the expected inverse proportionality between R_{s1} and the reverse current in different ambients is not apparent (if anything they seem to be directly proportional),

One way out of this difficulty is to consider a more complete equivalent circuit for the surface. Fig. 3 shows a schematic representation of a gold bonded diode, where $R_{\rm st}$ is the resistance along the surface which shunts the transition capacitance C_{l_1} and C_s is the capacitance associated with the surface space-charge region. In addition, we shall introduce a second surface resistance R_{s2} which expresses the dependence of the reverse current on the surface generation rate, and is analogous to the diffusion resistance at the bulk junction (omitted from the diagram because it is very large compared to the reactance of C_t). It is clear that R_{s2} is in parallel with C_s in order to provide a dc current path. Although Rst, Rs2, and C_s are represented as lumped constants, they are actually distributed over the surface and may vary from point to point.

Experimental information indicates that $C_i \gg C_i$, otherwise the measured equivalent

TABLE II CALCULATED LEAKAGE RESISTANCE IN DIFFERENT AMBIENTS

Ambient	Reverse Bias	R_{s1} (K Ohm)
Ozone	0 Volt	4.7
Ozone	-1	18
Dry O2	0	6
Dry O2	-1	47
Wet O2 Wet O2	0 -1	18 156



Fig. 3—(a) Schematic representation of gold-bonded diode. (b) Equivalent circuit of gold-bonded diode.

capacitance would differ from the transition capacitance C_t . At microwave frequencies, C_{\bullet} short-circuits $R_{\bullet 2}$ so that the equivalent series resistance is still expressed by (2). However, at dc and low frequencies, the main contribution to the dynamic barrier resistance is given by R_{s2} which may be much larger than $R_{\rm sl}$.

While this work was in progress, D. E. Sawyer of Lincoln Laboratories, Lexington, Mass., reported a 1/f frequency dependence (in the range 10 to 250 mc) for the equivalent series resistance of a variable capacity diode.² While our experimental conditions differ from his in several important instances (different diode structure, higher frequency range, etc.,) there is nevertheless no adequate explanation for this difference.

S. T. ENG R. Solomon Semiconductor Div. Hughes Products Newport Beach, Calif.

⁹ D. E. Sawyer, Device Research Conf., Ithaca, N. Y., June, 1959.

WWV Standard Frequency Transmissions*

Since October 9, 1957, the National Bureau of Standards radio stations WWV and WWVH have been maintained as constant as possible with respect to atomic frequency standards maintained and operated by the Boulder Laboratories, National Bureau of Standards. On October 9, 1957, the U.S.A. Frequency Standard was 1.4 parts in 109 high with respect to the frequency derived from the UT 2 second (provisional value) as determined by the U.S. Naval Observatory, The atomic frequency standards remain constant and are known to be constant to 1 part in 109 or better. The broadcast frequency can be further corrected with respect to the U.S.A. Frequency Standard, as indicated in the table; values are given as parts in 1010, This correction is not with respect to the current value of frequency based on UT 2. A minus sign indicates that the broadcast frequency was low.

The WWV and WWVH time signals are synchronized; however, they may gradually depart from UT 2 (mean solar time corrected for polar variation and annual fluctuation in the rotation of the earth). Corrections are determined and published by the U. S. Naval Observatory.

WWV and WWVH time signals are maintained in close agreement with UT 2 by making step adjustments in time of precisely plus or minus twenty milliseconds on Wednesdays at 1900 UT when necessary; a retarding time adjustment was made at WWW and WWVII on December 16, 1959.

WWV FREQUENCY WITH RESPECT TO U. S. FREQUENCY STANDARD

1959 1600 UT	Parts in 1010†
December 1	-30
2	-29
3	-29
4	-29
5	- 29
6	- 28
7	-28
8	- 28
9	-28
10	-28
11	-28
12	-28
1.5	-28
14	- 28
10	- 28
17	- 28
18	- 20
10	- 28
20	- 28
21	-28
22	-28
23	-28
24	-28
25	-28
26	-28
27	-28
28	-28
29	-28
30	- 28
31	-28

† 30-day moving average seconds pulses at 15 nc. Method of averaging is such that an adjustment of frequency of the control oscillator appears on the day it is made. No change or adjustment in the control oscillator was made during December, 1959. *Note:* Beginning January 1, 1960, the value of the USFS has been arbitrarily increased by 74 parts in 10¹⁰ to bring it into agreement with a cesium resonator frequency of 9192, 631, 770 cps. See "National stand-ards of time and frequency in the United States," National Burean of Standards, PRoc. IRE, vol. 48. pp. 105-106; January, 1960.

NATIONAL BUREAU OF STANDARDS Boulder, Colo.

The Tunnel-Emission Amplifier*

During the recent tumult caused by the "tunnel diode," this author had cause to reflect upon just what significant statements might be made concerning this device. The more important conclusions may be summarized as follows:

- 1) The device employs a *controlled source* of majority carriers.
- Its frequency response is essentially limited by the number of majority carriers available.
- In times past many negative resist-ance devices have been introduced, but in the course of time have given way to stable three-terminal amplifying devices.

Once interest in a negative resistance is abandoned, it becomes clear that semiconductors are of questionable value, since their carrier densities are inherently quite low. Metals with very large electron densities may be used as a carrier source, and insulators provide the necessary forbidden regions.

* Received by the IRE, December 28, 1959.

^{*} Received by the IRE, January 25, 1960.



Fig. 5—Asymptotic maximum gain-baudwidth prod-ucts for three connections of maximally-flat negative-conductance amplifiers vs number of poles. For the reflection type, $g_G = 1$. For the transmission type, $g_G = 1/(2n-1)$.



3. 6—Basic experimental tunnel (Esaki) diode amplifier. The dc biasing circuit was tailored to the i-v characteristics of one particular Esaki diode to permit fine control of setting to the point of maximum negative conductance. Fig. 6-



Fig. 7—Typical reflection-type maximally-flat 3-mc amplifier (n=2). All capacitors were provided with trimming adjustments to obtain the desired maximally-flat response, as observed on an oscilloscope display of gain vs frequency.

signed for $K_0 = 100$, $g_G = 5$, $G_D = (8.39)(10)^{-2}$ mho, and C_D (diode plus tuning capacitance) = 530 $\mu\mu f$. The measured bandwidths of various 3-mc reflection and 1-mc transmission-type amplifiers were in good agreement with theoretical values. (See Fig. 4 for the reflection-type results.) The selected value of $g_0 = 5$ results in a negligible load noise contribution, and thus the measured noise factor of 7.5 db was in good agreement with the theoretical shot noise term³ for the particular diode used.

The author acknowledges the help of A. Barone in measurements, and helpful discussions with J. C. Greene in preparing this letter.

> E. W. SARD Airborne Instruments Lab. Div. of Cutler-Hammer, Inc. Melville, N. Y.

^{*} K. K. N. Chang, "Low-noise tunnel-diode ampli-fier," Proc. IRE, vol. 47, pp. 1268-1269; July, 1959.

Frequency Dependence of the **Equivalent Series Resistance** for a Germanium Parametric **Amplifier Diode***

The equivalent circuit of a zero- or backbiased germanium diode is usually represented as a parallel combination of a transition region capacitance and a dynamic barrier resistance in series with a bulk resistance. From ac junction theory and skin affect considerations, these elements would be expected to be frequency independent to frequencies beyond 1 kmc. Early measurements of the equivalent series resistance (r_s') of germanium parametric amplifier diodes¹ indicated a frequency dependence in the range from 200 mc to 1 kmc. Furthermore, there were indications that r_s' was surface dependent. The present paper reports on a series of measurements performed in different ambients, and offers a model to explain the results.

The equivalent series resistance and capacitance was measured by exploration of the standing wave pattern in a slotted line. With a sensitive receiver as a detector, standing-wave ratios of around 500 could be measured. In order to get the true junction impedance Z_L of the diode from the measured admittance Yin, the following transformation was used:

$$Z_L = \frac{1}{Y_{\rm in} - Y_0} - \frac{1}{j(B_{\rm sh} - B_0)} \qquad (1)$$

where $Y_0 = G_0 + jB_0$, and $Y_{sh} = G_{sh} + jB_{sh}$ are the open-circuited and short-circuited diode admittances. The values of Z_L are calculated on a computer from the measured standing wave ratios and shift in minimum. The above transformation has been checked against a general four-terminal transformation and good agreement has been obtained. It is estimated that the measurement is accurate to 5 per cent at around 1 kmc and may be as high as 10 per cent at lower frequencies.

The measured values of equivalent series capacitance stayed constant to within ± 0.1 $\mu\mu$ fd with changing frequency, and the variation with bias was the same in the microwave region as in the low-frequency range, *i.e.*, $C_s' = C_0 (1 - V/\phi)^{-1/2}$ where ϕ is the diffusion potential and V is the bias voltage. Furthermore, C_s' was not affected by the changing ambients.

The equivalent series resistance r_s' for a typical diode is shown in Figs. 1 and 2 as a function of frequency for several different ambients. The curves drawn through the experimental points were obtained by fitting a $1/f^2$ law to the points.

Although measurements were made at higher reverse bias, they are not plotted because of the larger errors inherent in the measurement at high reverse bias. It can be said, however, that r_s' appeared to be relatively independent of applied bias.

* Received by the IRE, October 5, 1959. ¹ The measured diodes are Hughes gold-bonded germanium diodes available commercially as the HPA-2800 series. The properties of this diode are discussed in a paper hy S. T. Eng and W. Waters, presented at the Natl. Electronics Conf., Chicago, HI., October, 1959; to be published in *Proc. NEC*.

In general, r_s' was highest in ozone and lowest in wet oxygen. On the other hand, the dc reverse current is lowest in dry O2 and higher in both ozone and wet oxygen (see Table D.

The results can be explained by assuming a constant surface resistance shunting the transition capacitance C_t . Since the Q of the diode is high, the equivalent series resistance for such a circuit is given by

$$r_{s}' = r_{b} + \frac{1}{\omega^{2} C_{l}^{2} R_{s1}}$$
(2)

where r_b is the bulk resistance and R_{s1} is the surface shunt resistance. Eq. (2) has been been fit to the data, and it is indicated by the curves drawn through the experimental points in Figs. 1 and 2. It is seen that r_s' in ozone and dry oxygen fits (2) quite well,



Fig. 1—Frequency dependence of equivalent series resistance, 0 volt bias.



Fig. 2—Frequency dependence of equivalent series resistance. -1 volt bias.

TABLE 1 REVERSE CURRENTS IN DIFFERENT AMBIENTS

Ambient	Reverse Bias	I_r (mµamp)
Ozone	-1 Volt	150
Ozone	-2	260
Ozone	-3	440
Dry O ₂	-1	90
Dry Or	-2	150
Dry O ₂	-3	320
Wet O ₂	-1	300
Wet O ₂	-2	680
Wet O ₂	-3	1550

1960

Correspondence_

Tunnel (Esaki) Diode Amplifiers with Unusually Large Bandwidths*

The bandwidth of an amplifier whose essential element is a frequency-invariant negative conductance in shunt with a parasitic capacitance, for example the tunnel (Esaki) diode,1 is limited and tends to decrease with increasing gain. In particular, if the parasitic capacitance is resonated in a single-tuned circuit, the product of the square root of transducer-power gain and bandwidth is almost constant at high gain. Recently it was shown that for the degenerate parametric amplifier, another form of negative-conductance amplifier, maximallyflat filter circuits can be used instead of a single-tuned circuit in order to greatly increase the bandwidth.2 The invariant gainbandwidth product then becomes (power gain)^{1/2n} times bandwidth, where n is the number of poles in the filter. The general analysis of Seidel and Herrmann² has been extended; a summary of this analysis, together with corroborating experimental data is presented here for three connections of negative-conductance amplifiers operated with lumped-constant filters that give an over-all maximally-flat response. A complete account of this analysis will be published at a later date.

One connection, denoted reflection type (see Figs. 1 and 2), has the generator and load located at one port of the filter, and the negative conductance at the other port. Filter-element values for a ladder structure can be chosen to give a maximally-flat (but nonminimum phase) response. Design equations for the cases of n = 2 and n = 3 are given in Fig. 3. Note that the element values are functions of the prescribed gain. Fig. 4 shows the very large increases in bandwidth predicted for this connection, compared with those obtained using single-tuned circuits, particularly at high gain. Note that, at a given gain, the principal relative increase in bandwidth is in going from n=1 to n=2. The n=3 filter does have the advantage, however, of permitting parasitic shunt capacitance in the load circuit to be incorporated in the filter.

Another connection is the reflection type operated with a circulator to isolate the generator and load. This is the configuration previously described,² and it has the usual advantages of operating a negative conductance amplifier with a circulatorthat is, greater bandwidths for the same gain and greater gain stability against changes in generator and load values. The response is similar to that for operation without a circulator; and design equations can be related to those in Figs. 2 and 3 as follows The new value of h is

July, 1959. ² H. Seidel and G. F. Herrmann, "Circuit aspects of parametric amplifiers," 1959 IRE WESCON CON-VENTION RECORD, pt. 2, pp. 83–90.



- 1—Low-pass reflection-type maximally-flat negative-conductance amplifier. For band pass, insert a capacitance in series with each series inductance and an inductance in shurt such sciences shurt capacitance, each pair resonating at the center frequency. (a) Low-pass circuit; (b) normal-ized low-pass circuit.

$$h = \frac{G_{G} + G_{L}}{G_{D}} > I \qquad g_{G} = \frac{G_{G}}{G_{L}}$$

$$C_{I} = \frac{C_{D}}{G_{D}} \qquad Z = \frac{G_{D} V_{I}}{I_{I}}$$
POWER GAIN, $K = \rho h^{2} \left| \frac{Z}{I + hZ} \right|^{2}, \ \rho = \frac{4g_{G}}{\left(1 + g_{G}\right)^{2}}$

$$K_{O} = \frac{\rho h^{2}}{\left(h - I\right)^{2}} \quad OR \quad h = \frac{1}{I - \sqrt{\frac{\rho}{K_{O}}}}$$

Fig. 2—Basic equations for Fig. 1. The parameters of the normalized Fig. 1(b) are related to those of Fig. 1(a). Note that the parasitic capacitance of the negative conductance is incorporated in the filter.

$$h' = \frac{G_0}{G_D} > 1 \tag{1}$$

where G_0 is the characteristic conductance of the circulator. The new values of power gain are

$$K' = \left|\frac{h'Z - 1}{h'Z + 1}\right|^2 \tag{2a}$$

$$K_{0}' = \left(\frac{h'+1}{h'-1}\right)^2$$
, or $h' = \frac{\sqrt{K_0'}+1}{\sqrt{K_0'}-1}$. (2b)

The parameter *h* in the equations for α and β is replaced by the quantity $[1+(h')^2]/2$; or, alternatively, these equations hold for a new gain, $K_0' = [\sqrt{K_0/\rho} + \sqrt{(K_0/\rho) - 1}]^2$. The normalized bandwidth equations, however, take a different form:

$$\left(\frac{\omega_{C'}C_{D}}{G_{D}}\right)^{2} = \left(\frac{h'+1}{\alpha'h'}\right) \left(\frac{1}{\sqrt{K_{0}'-2}}\right),$$

$$n = 2 \quad (3a)$$

$$\left(\frac{\omega_{C'}C_{D}}{G_{D}}\right)^{3} = \left(\frac{h'+1}{3\beta'-1}\right) \left(\frac{1}{\sqrt{K_{0}'-2}}\right),$$

$$n = 3. \quad (3b)$$

Although the actual bandwidths are larger than for operation without a circulator, the relative increases for increasing values of nare somewhat smaller.

The third connection, denoted transmission type, has the generator and negative conductance located at one port of the filter. and the load at the other port. Again,

$$S_{L} = \frac{G_{D}}{1 + q_{0}^{2} - 2\left[\frac{q_{0}^{2}}{K_{0}} + G_{0}^{2} + q_{0}^{2}G_{L}^{2}\right]}$$

$$= 2 \cdot L_{A} = \frac{aC_{D}}{G_{0}^{2}} = \frac{1 - \sqrt{\frac{2(h-1)}{2h-1}}}{\frac{2(h-1)}{2h-1}}$$

$$= 3 \cdot L_{A} = \frac{aC_{D}}{G_{D}^{2}}, \quad G_{B} = \beta C_{D}, \quad a = 3 - \frac{1}{\beta}$$

$$= 3 \cdot C_{A} = \frac{aC_{D}}{G_{D}^{2}}, \quad G_{B} = \beta C_{D}, \quad a = 3 - \frac{1}{\beta}$$

$$= 3 \cdot C_{A} = \frac{aC_{D}}{G_{D}^{2}}, \quad G_{B} = \beta C_{D}, \quad a = 3 - \frac{1}{\beta}$$

 $\frac{\left(C_{D}\right)^{2}}{D} ROOT OF, \frac{4(3\beta-1)^{4}}{(1-\beta)^{6}} X^{3} - \left(3 - \frac{1}{\beta}\right)^{2} X^{2} + \left(3 - \frac{1}{\beta}\right)\left(\frac{1}{\beta} - 1\right) X - I = 0$

z. 3—Design equations for reflection type. The quantities g_G , K_0 , G_D , and C_D are assumed specified. L_A is the series inductance and C_B is the second shunt capacitance in the low-pass filter. Fig.



-Normalized bandwidths vs power gain for Fig. 4-(*n*=1) amplifiers. The angular bandwidth (ω_c) is normalized to the figure of merit of the negative conductance, Gn/Cn.

(generally different) filter-element values for a ladder structure can be chosen to give a maximally-flat (and minimum phase) response. In general, the transmission type has less bandwidth than the reflection type, the difference becoming greater for ratios of source-to-load conductance that give lower noise factors.

Now, consider the asymptotic maximum gain-bandwidth performance at high gain for the three connections (Fig. 5). It can be shown that the curves for all three connections approach the same decreasing curve as n approaches infinity. Furthermore, for the reflection type operated with a circulator, the ordinate in Fig. 5 equals 2 sin $(\pi/2n)$. Thus, the gain-bandwidth products in Fig. 5 approach zero as n approaches infinity, and (contrary to Seidel and Herrmann) it is not theoretically possible to operate in the limit with nonzero bandwidth and infinite gain. Instead, there will be some value of n that maximizes the gain for a given bandwidth.

The basic experimental amplifier (Fig. 6) was a band-pass design. Fig. 7 shows a typical reflection-type maximally-flat 3-mc amplifier (n=2). This amplifier was de-

^{*} Received by the IRE, December 28, 1959. The work reported here was performed under Air Force Contract AF30(602)-1854. ¹ H. S. Sommers, Jr., "Tunnel diodes as high fre-quency devices," PROC. IRE, vol. 47, pp. 1201-1206; July, 1959.



Fig. 10—Variation of plasma frequency $(f_p = 1/2\pi \sqrt{ne^2/n\epsilon_0})$ and electron collision frequency ν at the stagnation point of a hypersonic vehicle with velocity at various altitudes above the earth.



Fig. 11—Electromagnetic properties of air at the stagnation point of a hypersonic vehicle in the atmosphere at frequencies of 3×10^9 and 30×10^9 cps, showing the variation of the attenuation and phase constants with altitude and velocity.

Acknowledgment

The authors are indebted to the Aerophysics Wing of the Canadian Armament Research and Development Establishment for financial support.



Fig. 9—Electromagnetic properties of high-temperature air at frequency of 6×10⁹ cps, showing the variation of the attenuation and phase constants at constant temperature, constant density, and constant reflection coefficient.

plasma behaves nearly like free space at all densities. As the temperature increases, the influence of the ambient density becomes more apparent as the plasma becomes more lossy. At high temperatures, the plasma is a good conducting medium and the effect of density variation again becomes secondary. In this representation, an "operating region" for propagation of electromagnetic energy can be determined, provided the tolerable attenuation and reflection coefficients are specified.

Hypersonic Vehicle in Planetary Atmosphere

A space vehicle moving at hypersonic velocities within a planetary atmosphere will be surrounded by a shock-induced ionized sheath. If the temperature and density of a given region of the shock is known, then the results presented earlier (Figs. 4–8) can be used to obtain an estimate of the propagation characteristics of an electromagnetic wave through this region. For purposes of communicating to and from a space vehicle, it is most feasible to attempt to propagate electromagnetic energy through the wake of the shock or at some point aft of the stagnation region of the shock where the influence of the plasma on an electromagnetic wave is not as pronounced as in the stagnation region; *i.e.*, the electron density and collision effects are less. However, one difficulty is that the temperature and density distribution of the shock away from the stagnation region is not very well known. On the other hand, it may be instructive to determine the propagation characteristics of an electromagnetic wave in the stagnation region of a hypersonic vehicle for two reasons, firstly that assuming thermal equilibrium the thermodynamic quantities in this region can be readily deduced from aerodynamic considerations, and, secondly, that this is the region of most dense plasma and hence the most critical conditions for penetration by, or propagation of, an electromagnetic wave.

Fig. 10 shows the variation of plasmå frequency and electron collision frequency at the stagnation point of a hypersonic vehicle with velocity at various altitudes above the earth. (The ARDC model of the atmosphere was used in the aerodynamic considerations.¹³) Using the values in Fig. 10, contours of constant velocity and constant altitude for a hypersonic body have been plotted in the normalized propagation plane for frequencies of 3 kmc and 30 kmc. These are shown in Fig. 11.

¹³ A. R. Hochstim and R. J. Arave, "Various Thermodynamic Properties of Air," Convair, San Diego, Calif., Rept. No. ZPH-004; June, 1957.

For high densities, the collision terms (*i.e.* imaginary part of the dielectric constant) predominate and no pronounced minimum values occur.

The frequency dependence of the normalized attenuation and phase constants are illustrated in Figs. 7 and 8. At a temperature of 3000°K, air is not sufficiently ionized to appreciably affect the propagation characteristics of an electromagnetic wave except for low frequencies. (The present discussion is confined to frequencies above 1 kmc, although similar analyses can be applied to the lower frequencies using the methods and data presented earlier.) At a temperature of 3000°K, air acts like a slightly lossy dielectric whose dielectric constant is nearly unity. Consequently, the phase constant is essentially the same as for free-space propagation and the attenuation constant is very small. Air in thermal equilibrium at a temperature of 6000°K can act as either a dielectric or a conductor, depending on the density and RF frequency. At low densities and high RF frequencies, air is essentially a dielectric with $\beta/k\sim 1$, and α/k is small. However, as the RF frequency is decreased, a rapid rise in attenuation and change in phase constant occurs as the plasma frequency becomes comparable to or greater than the frequency of the impressed electromagnetic wave. At high densities of air, the attenuation and phase constants are in general quite large, and decrease with increasing RF frequency. A-12,000°K, air is essentially a good conductor, exhibiting very high attenuation and phase characteristics.

The propagation characteristics of high temperature air are represented in the propagation plane for an impressed frequency of 6 kmc in Fig. 9. Contours of constant temperature, constant density, and constant reflection coefficient are shown. At low temperatures, the



Fig. 7—Dependence of attenuation constant (α/k) on frequency for high-temperature air. $(\rho_0 = 1.28823 \times 10^{-3} \text{g/cm}^3.)$



Fig. 8—Dependence of phase constant (β/k) on frequency for high-temperature air. $(\rho_0 = 1.28823 \times 10^{-3} \text{g/cm}^3.)$



Fig. 5—Dependence of attenuation constant (α/k) on temperature and density of air. ($\rho_0 = 1.28823 \times 10^{-3} g/cm^3$.)



Fig. 6—Dependence of phase constant (β/k) on temperature and density of air. $(\rho_0 = 1.28823 \times 10^{-3} \text{g/cm}^3.)$

The normalized attenuation and phase constants for air in thermal equilibrium are shown as a function of temperature and density for RF frequencies of 1, 10, and 100 kmc in Figs. 5 and 6, respectively. The attenuation constant α/k increases with increasing temperature as the number of electrons and collisions become greater. Values of α/k decrease with increasing RF frequency; however, this does not mean that the attenuation necessarily decreases, since k is increasing with increasing frequency. The phase constant β/k starts off from its free-space value of unity at low temperatures, and for low densities at first decreases with increasing temperature and then increases with temperature passing through a minimum in the region where $\omega_p^2 \sim \omega^2$.

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Fig. 4—Variation of electron density (*n*) and electron collision frequency (ν) with temperature for air at different densities. ($\rho_0 = 1.28823 \times 10^{-3} \text{ g/cm}^3$.)

this paper. The method is outlined below. Recent, more rigorous work¹⁰ indicates that in the range 6000°K to 12,000°K, the total collision frequency is nearly a constant with velocity, so that (36) can be used but the values are slightly different from Fig. 4.

The method of constant mean free path¹¹ which

¹⁰ I. P. Shkarofsky, M. P. Bachynski, and T. W. Johnston, "Collision Frequency Associated with High Temperature Air and Scattering Cross-Sections of the Constituents," presented at AFCRC Symposium on the Plasma Sheath, Boston, Mass.; December 7–9, 1959.

¹¹ L. G. H. Huxley, "Free path formulas for the electronic conductivity of a weakly ionized gas in the presence of a uniform and constant magnetic field and a sinusoidal electric field," *Aus. J. Phys.*, vol. 10, pp. 240–245; 1957. assumes the mean free path of the electrons to be independent of electron velocity is used here to calculate the collision frequency. Making this assumption, the collision frequency for a Maxwellian distribution of electron velocities at high RF frequencies becomes⁵

$$\nu = \bar{c} \sum n_j (\partial_j, \qquad (16a)$$

where

- n_j = number density of the *j*th specie,
- $Q_j =$ Maxwell-averaged total electron collision cross section of the specie,

$$\tilde{z} = \frac{4}{3}$$
 times mean electron velocity $= \frac{4}{3} \sqrt{\frac{8\kappa T}{\pi m}}$, (16b)

 $\kappa = \text{Boltzmann's constant},$

T = temperature (°K).

With the collision cross sections of the neutral species as given by Massey and Burhop¹² and calculating the conductivity due to electron-ion collisions^{7,8} from

$$\sigma_{\rm ion} = \frac{1.1632m}{\ln (h/b_{\rm u})} \left(\frac{4\pi\epsilon_0}{e}\right)^2 \left(\frac{2kT}{\pi m}\right)^{3/2}$$

where

$$h = \left(\frac{\kappa T 4 \pi \epsilon_0}{8 \pi n e^2}\right)^{1/2}; \qquad b_0 = \frac{e^2}{(4 \pi \epsilon_0)(3 \kappa T)}$$

The variation of collision frequency with temperature and density have been evaluated using (16a). These are shown in Fig. 4 for temperatures ranging from 3000° to $12,000^{\circ}$ K and densities from 10^{1} to 10^{-4} times standard density. Although the collision cross sections of the neutral species may not be accurate, the values fall within the spread of estimates which can be obtained from the current literature. Further, at the higher temperatures the positive ion effects predominate and the influence of the neutral species becomes of lesser and lesser importance. The above plots are thus indicative of the expected variation and more accurate computation must await further experimental determination of scattering cross sections and further refinements in theory.¹⁰

Attenuation and Phase Constants

From the values of the electron density and collision frequency shown in Fig. 4, the attenuation and phase constants of an electromagnetic wave propagating in the plasma created by air at high temperatures can be determined by using (4a), (4b), (6c), (6d), and (6e).

¹² H. S. W. Massey and E. H. S. Burhop, "Electronic and Ionic Phenomena," Oxford University Press, New York, N.Y.; 1952. dent electromagnetic wave reflected at a plasma-free space interface is given by the reflection coefficient Rwhere from (6)

$$R = \frac{Z - Z_0}{Z + Z_0} = \frac{1 - K^{1/2}}{1 + K^{1/2}}$$
(14)

$$= \frac{(1-B)+j.4}{(1+B)-jA} .$$
(14a)

The magnitude of the reflection coefficient is given by

$$|R| = \sqrt{\frac{1+A^2+B^2-2B}{1+A^2+B^2+2B}}$$
(15a)

$$= \sqrt{\frac{1-x}{1+x}}.$$
 (15b)

where

$$x = \frac{2B}{(1 + A^2 + B^2)}.$$
 (15c)

Similarly, the magnitude of the transmission coefficient is

$$|T| = \sqrt{1 - |R|^2} = \sqrt{\frac{2x}{1 + x}}$$
 (15d)

For any magnitude of reflection coefficient, the relationship between the attenuation A and the phase B is uniquely determined by (15c), which can be rewritten

$$A^{2} + (B - 1/x)^{2} = \left(\frac{1}{x^{2}} - 1\right),$$
 (15d)

which is a family of circles with center (A = 0, B = 1/x)and radius $[(1-x^2)/x^2]^{1/2}$.

Families of constant reflection coefficient plotted in the propagation plane are shown in Fig. 3. If in a par-





ticular application the maximum allowable attenuation and reflection coefficient are known, then the "operating region" in the propagation plane is defined by the area enclosed by the *B*-axis and the lines $B^2 - A^2 = 1$, $R = R_{\text{max}}$, and $\alpha/k = \alpha_{\text{max}}/k$. Similar plots and operating regions may be determined for the transmission properties $T_{\min} = \sqrt{1 - |R_{\max}|^2}$ of a plasma.

ELECTROMAGNETIC PROPERTIES OF HIGH TEMPERATURE AIR

Electron Density and Collision Frequency

At chemical equilibrium, the temperature of a hot gas has reached a stable value so that all the constituents within the plasma have been brought to this equilibrium temperature. The molar fraction of these constituents. as well as all other thermodynamic quantities, can be calculated by the methods of quantum statistical thermodynamics4,5 which determine the partition of the energy states of a particle species into translational. rotational, vibrational, and electronic energies as well as the energies of dissociation and ionization. Tables are available giving the equilibrium quantities for air.

The electron concentration in high temperature air in thermal equilibrium as a function of temperature and density as determined by Gilmore⁶ is shown in Fig. 4. At temperatures above 3000°K a rapid increase in the number of electrons occurs with rise in temperature as ionization takes place. This rate of increase gradually levels off at the higher temperatures as all the constituents become singly ionized and the number of electrons does not increase substantially with temperature.

The collision frequency of the electrons in a high temperature gas mixture such as air is generally calculated by adding the Maxwell-averaged collision cross section (weighted according to the number density of each species) of the neutral molecules to a corresponding equivalent cross section of the ions, the electron-ion collision cross section being calculated according to Spitzers7 theory for fully ionized gases.8,9 This method was thought to be a better approximation than the mean-free-time (constant γ) method and was used for

⁴ J. G. Logan, Jr., "The Calculation of the Thermodynamic Properties of Air at High Temperatures," Cornell Aeronautical Lab., Inc., Ithaca, N. Y., Rept. No. AD-1052-A.1; May, 1956.
⁶ M. P. Bachynski, I. P. Shkarofsky, and T. W. Johnston, "Plasma Physics of Shock Fronts," Res. Labs., RCA Victor Co., Ltd., Montreal, Can. RCA Res. Rept. No. 7-801-3; June, 1959.
⁶ F. P. Cilaren, "Feuilibrium Convestion and Thermodynamic Conversion and Co

⁶ F. R. Gilmore, "Equilibrium Composition and Thermodynamic Properties of Air to 24,000°K," RAND Corp., Santa Monica, Calif., Rept. No. RM-1543; 1955. 7 L. Spitzer and R. Harm, "Transport phenomena in a completely

ionized gas," *Phys. Rev.*, vol. 89, pp. 977-81; 1953. * L. Lamb and S. C. Lin, "Electrical conductivity of thermally ionized air produced in a shock tube," *J. Appl. Phys.*, vol. 28, pp. 754-759; July, 1957.

⁹ M. P. Bachynski, I. P. Shkarofsky, and T. W. Johnston, "Plasmas and the Electromagnetic Field," Res. Labs., RCA Victor Co., Ltd., Montreal, Can., RCA Res.Rept. No. 7-801-2; November, 1958.

The normalized collision parameter C can be expressed in terms of A and B by the use of (12a) and (13a) in (9a). Similarly, the values of normalized frequency F in the propagation plane is obtained by substituting (13a) into (10).

Contours of constants S and N in the complex propagation constant plane are shown in Fig. 2(a), and plots of constants C and F in Fig. 2(b). Extended values of these parameters are shown plotted on a logarithmic scale in Figs. 2(c) and 2(d).

It will be observed that the normalized parameters map more readily into the complex dielectric coefficient plane than into the propagation constant plane. However, in measurement or diagnostics it is the attenuation and phase constants which are actually determined. Hence, the added difficulty in plotting the normalized contours in the propagation constant plane is generally justified.

A second reason for representing the plasma parameters in the propagation plane is that the reflecting or transmitting properties of a plasma boundary are very conveniently mapped in the propagation plane. Thus, if Z is the impedance of the plasma and Z_0 the impedance of free space, the fraction of the field of a normally inci-



Fig. 2—(a) Variation of normalized scattering frequency (S) and normalized electron density (N) in complex propagation constant plane. $(S = \nu/\omega; N = (\omega_p/\omega)^2)$ ($\nu =$ electron collision frequency; $\omega =$ RF frequency, $\omega_p =$ plasma frequency.) (b) Variation of normalized collision parameter (C) and normalized RF frequency (F) in complex propagation constant plane. $(C = \nu/\omega_p; F = \omega/\omega_p)$ (c) Variation of normalized scattering frequency (S) and normalized electron density (N) in complex propagation constant plane for extended values of the parameters (logarithmic scale). $(S = \nu/\omega; N = (\omega_p/\omega)^2)$ (d) Variation of normalized collision parameter (C) and normalized RF frequency (F) in complex propagation constant plane for extended values of the parameters (logarithmic scale). $(C = \nu/\omega_p; F = \omega/\omega_p)$.

which is a family of straight lines in the complex dielectric (Kr vs Ki) plane, with slope -S and Kr-intercept of 1.

Similarly, the normalized electron density is

$$N = 1 - Kr + Ki^2/(1 - Kr)$$
, or (8a)

$$(Kr - (1 - N/2))^2 + Ki^2 = (N/2)^2,$$
 (8b)

which in the complex plane is a family of circles of radius N/2 and center (Kr = (1 - N/2), Ki = 0).

Representation in terms of the normalized collision parameter is slightly more difficult in that

$$C^2 = S^2/N, \tag{9a}$$

or in terms of the real and imaginary part of the dielectric coefficient,

$$(1 - Kr)^{3} + Ki^{2}(1 - Kr) - Ki^{2}/C^{2} = 0$$
 (9b)

or

$$Ki = (1 - Kr) \left(\frac{1}{C^2(1 - Kr)} - 1 \right)^{-1/2}, \qquad (9c)$$

with a pole at $Kr = 1 - 1/C^2$.

The normalized RF frequency loci are again circles in the complex dielectric coefficient plane since

$$F = 1/\sqrt{N} \tag{10}$$

The radius of the *F*-circles is $1/2F^2$ and their centers are located at $[Kr = (1 - 1/(2F^2)), Ki = 0]$.

Families of constant scattering frequency and constant electron density normalized to signal frequency plotted in the dielectric plane are shown in Fig. 1(a), while contours of constant collision parameter and RF frequency normalized to the plasma frequency are shown in Fig. 1(b).

Propagation Plane

Representation of the normalized parameters in the complex propagation constant plane can be derived from a conformal transformation of values in the dielectric plane, since from (6a), (6c) and (6d),

$$A + jB = jK^{1/2},$$
 (11)

or from a direct solution for the normalized parameters in terms of the attenuation and phase constants. Since the maximum value of the real part of the dielectric constant is unity, the upper limit in the propagation plane is given by the line Kr = 1 which maps into the hyperbola $B^2 - A^2 = 1$.

The normalized scattering frequency loci become rectangular hyperbolas rotated through an angle of $1/2 \tan^{-1}(1/S)$ in the complex propagation constant plane, namely

$$S = \frac{2AB}{1 - (B^2 - A^2)}$$
, or (12a)

$$B^2 - A^2 + 2AB/S = 1.$$
 (12b)

The normalized electron density families become quartic curves since

$$N = 1 + A^2 - B^2 + 4A^2B^2/(1 + A^2 - B^2)$$
, or (13a)

$$(A^{2} + B^{2})^{2} - (2 - N)(B^{2} - A^{2}) + (1 - N) = 0.$$
 (13b)



Fig. 1—(a) Variation of normalized scattering frequency (S) and normalized electron density (N) in the complex dielectric coefficient plane. $(S = \nu/\omega; N = (\omega_p/\omega)^2)$ ($\nu =$ electron collision frequency, $\omega =$ RF frequency, $\omega_p =$ plasma frequency.) (b) Variation of normalized collision parameter C and normalized RF frequency (F) in the complex dielectric coefficient plane. $(C = \nu/\omega_p; F = \omega/\omega_p)$

v is the electron velocity, and

 ν is the effective collision frequency of the electrons (equal to the reciprocal of the time between successive collisions).

If the collision frequency is independent of electron velocity, (3a) simplifies to

$$\sigma = \frac{ne^2}{m(\nu + j\omega)} \tag{3b}$$

where *n* is the number density of electrons, which is a reasonable approximation for air at temperatures considered here (particularly 6000° K-12,000°K).

Using (3b) in (2), the dielectric coefficient becomes

$$K = \left\{ 1 - (\omega_p/\omega)^2 \frac{1}{1 + (\nu/\omega)^2} \right\}$$
$$- j \left\{ (\omega_p/\omega)^2 \frac{(\nu/\omega)}{1 + (\nu/\omega)^2} \right\}$$
(4a)

$$= Kr + jKi \tag{4b}$$

where the parameter $ne^2/\epsilon_0 m = \omega_p^2$ has the dimensions of seconds⁻² and ω_p is the "plasma frequency." *Kr* and *Ki* are the real and imaginary parts, respectively, of the dielectric coefficient.

Propagation Constants

In a uniform plasma it is assumed that the electron density in the absence of a field is not a function of position and in a neutral plasma there is no net charge, so that the charge density is zero. Hence, the electric field in the plasma is divergenceless, *i.e.* $\nabla \cdot E = 0$. Further, by the usual manipulation of Maxwell's third and fourth equations the Helmholz vector equations for Eand H are obtained. Thus

$$\nabla^2 E + k^2 K E = 0$$

$$\nabla^2 H + k^2 K H = 0,$$
(5a)

where

 $k = \omega/c = 2\pi/\lambda$, c = velocity of light, $\lambda =$ free-space wavelength.

It is seen that both E and H satisfy the same vector equation. For any specific problem, solving either, subject to the proper boundary conditions, will give the complete solution.

The solutions of (5a) for a plane wave propagating through a uniform plasma are given in rectangular coordinates by

$$E_x = E_0 \epsilon^{i\omega i} \epsilon^{-\gamma z}, \qquad E_y = E_z = 0;$$

$$H_y = H_0 \epsilon^{i\omega i} \epsilon^{-\gamma z}, \qquad H_x = H_z = 0;$$
(5b)

where γ , the propagation constant, is defined by

$$\gamma = jkK^{1/2}.$$
 (6a)

 γ is complex, and of the two solutions for γ which differ in sign, the solution yielding a negative real part in the exponent is chosen in order to insure attenuation as the wave propagates. Thus

$$\gamma = \alpha + j\beta, \tag{6b}$$

where

$$\frac{\alpha}{k} = \sqrt{\frac{|K| - Kr}{2}} = A, \qquad (6c)$$

$$\frac{\beta}{k} = +\sqrt{\frac{|K| + Kr}{2}} = B, \tag{6d}$$

$$K = (Kr^2 + Ki^2)^{1/2}, (6e)$$

and α is called the attenuation constant, β the phase constant, of the plasma.

It is seen that the propagation constants of a plasma are determined by the effective dielectric coefficient which in turn depends on the electron density and collision frequency.

Universal Representation of Electromagnetic Parameters for Constant Collision Frequency

It is instructive to rewrite the relationships determining the complex dielectric coefficient and propagation constant in normalized form, and hence to demonstrate their behaviour in universal coordinates. Thus, define the following parameters

- Normalizing with respect to frequency: S = ν/ω = normalized scattering frequency, N = (ω_p/ω)² = normalized electron density.

 Normalizing with respect to plasma frequency
- (*i.e.* $n^{1/2}$): $C = \nu/\omega_p = \text{normalized collision parameter,}$

 $F = \omega / \omega_p = \text{normalized RF frequency.}$

These relationships permit the mapping of loci of constant scattering frequency S, constant electron density N, constant collision parameter C, or constant RF frequency F on the complex dielectric coefficient plane and on the complex propagation constant plane. S and Nare the useful parameters to consider in a diagnostic measurement, with a given frequency and varying plasma, while C and F are useful when the frequency behavior of a given plasma is of interest.

Dielectric Plane

Using (4a) for the complex dielectric coefficient, it is easily shown that the normalized scattering term is given by

$$S = Ki/(Kr - 1)$$
, or (7a)

$$(Kr - 1)S + Ki = 0,$$
 (7b)

Electromagnetic Properties of High-Temperature Air*

M. P. BACHYNSKI[†], T. W. JOHNSTON[†], and I. P. SHKAROFSKY[†]

Summary-This paper concerns the attenuation and phase characteristics of plasmas and, in particular, the electromagnetic properties of high-temperature air. It is shown that by a suitable normalization of the parameters the electromagnetic properties of plasmas may be universally represented in convenient form in either the complex dielectric coefficient plane or the complex propagation constant plane. Next, the electron number densities and electron collision frequencies for air ranging in temperature from 3000° to 12,000°K and in density from 10^1 to 10^{-4} times the density at sea level are illustrated. The attenuation and phase constants for an electromagnetic wave traversing this medium have been evaluated for frequencies from 10^9 to 10^{11} cps. As an example, the above universal representation is applied to the stagnation region of a hypersonic vehicle in space.

INTRODUCTION

UE to ionization, air at high temperatures contains an appreciable number of free electrons and ions. Under these conditions, the medium may be described as a plasma, *i.e.*, a gas containing charged particles in a sufficient quantity to seriously alter the physical properties of the gas. One of the properties of air markedly affected by the presence of the electrons and ions is the propagation of electromagnetic waves in such a medium. This interaction of electromagnetic waves with plasmas is of current interest in connection with diagnostic techniques, space communications, and re-entry problems.

The following paper is concerned with the electromagnetic characteristics of plasmas and, in particular, those of high temperature air. It is shown that by a suitable normalization of parameters, these properties can be represented in a convenient, universal form in either the complex dielectric coefficient plane or the complex propagation constant plane. Values of the electron density and electron collision frequency are shown for air in the temperature range 3000° to 12,000° K, and densities ranging from 10^{1} to 10^{-4} times the density at sea level. Further, the attenuation and phase constants of electromagnetic wave propagating in a medium of air at high temperatures are evaluated for radio frequencies ranging from 10⁹ to 10¹¹ cps.

Finally, as an example, the variation of attenuation and phase of an electromagnetic wave with altitude and velocity is determined for the stagnation region of a hypersonic vehicle.

ELECTROMAGNETIC PARAMETERS OF A UNIFORM PLASMA¹

Dielectric Coefficient

The dielectric coefficient of an infinite uniform plasma, *i.e.*, a plasma where electron density is not a function of position in the absence of an electromagnetic field, can be deduced from Maxwell's fourth equation. Thus, assuming a harmonic field variation $e^{j\omega t}$ and the permittivity of the plasma to be the same as free space permittivity ϵ_0 , one can write, using rationalized mks units.

$$\nabla \times \overrightarrow{H} = \overrightarrow{J} + \frac{\partial}{\partial t} (\epsilon_0 \overrightarrow{E})$$
(1a)

$$= \sigma \vec{E} + j\omega\epsilon_0 \vec{E}$$
(1b)

$$= j\omega\epsilon_0 K \overrightarrow{E}, \qquad (1c)$$

where

H and E are the magnetic and electric fields respectively of an impressed electromagnetic wave incident on the plasma,

J is the ac current density.

 $\epsilon_0 E$ represents the electric displacement,

 σ is the ac electronic conductivity of the plasma,

 ω is the radian frequency of the electromagnetic wave, $j = \sqrt{-1}$, and

K is the effective dielectric coefficient given by

$$K = 1 + \frac{\sigma}{j\omega\epsilon_0} \,. \tag{2}$$

In the absence of a dc magnetic field, the electronic conductivity σ of a plasma to an RF signal of frequency ω is given by^{2,3}

$$\sigma = -\frac{4\pi}{3} \frac{e^2}{m} \int_0^\infty \frac{1}{\nu + j\omega} \frac{\partial f_0^0}{\partial v} v^3 dv \qquad (3a)$$

where

e and m are the electronic charge and mass respectively.

 f_0^0 is the electron velocity distribution function,

^{*} Original manuscript received by the IRE, August 5, 1959. This paper was presented at the URSI International Symposium on Elec-tromagnetic Theory, Toronto, Can., June 15–20, 1959. † Res. Labs., RCA Victor Co., Ltd. Montreal, Canada.

¹ This section is intended only as a summary to define the various parameters.

² W. P. Allis, "Motion of ions and electrons," *Handbuch der Physik* vol. 21, Springer-Verlag, Berlin; 1956.
³ H. Margenau, "Conductivity of plasmas to microwaves," *Phys. Rev.*, vol. 109, pp. 6–9; January, 1958.





ionogram) the velocity with which the ray passes through the lower parts of the ionosphere approaches the light velocity. The traveling-time curves (Fig. 9) demonstrate this fact. The ionograms are drawn with a linear frequency scale and a traveling-time scale much larger than normally in use. This facilitates the analysis of the ionogram. A paper on this subject is in preparation.

Figs. 10 and 11 give oblique incidence curves for 1600 km distance for models 14, 20, and 12 as well as for the Lindau-Helsinki experiments [7]. Also, for this sort of experiment models may be found which fit the situation. It will be clear to the reader that the "horizontal" lines in Figs. 10 and 11 are a demonstration of the aforementioned fact, that the horizontal velocity is strongly independent of frequency and slightly dependent on the radiation angle (see Fig. 4).

VI. OTHER METHODS

The fundamental problem for ionospheric propagation studies is to find the distribution of electrons in the ionosphere and to deduce this distribution from ionospheric sounding. A very good survey of this problem was given by Thomas [6].

Formulas giving the amount of electrons in a vertical cylinder up to a variable height for each ionospheric model have been published elsewhere [10]. These values, which have not yet been produced in the form of curves, can also be obtained by simple planimetry below the ionization curves of each model down to zero N.

VII. CONCLUSION

The method of studying ionospheric models is suggested as a way of attacking the problem of ionospheric



Fig. 11-Oblique incidence sweeping frequency, Lindau-Helsinki.

radio propagation. The ionograms, which are available in great quantity nowadays, yield E and F critical frequencies, but the conventional methods, by which the practical values of MUF, etc., are produced, are not very accurate. The method described above might give a more sound basis for frequency-predictions.

VIII. ACKNOWLEDGMENT

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- Section Ionosphere and Radio Astronomy, Netherlands Posts [10] and Telecommunications Services, The Hague, Netherlands, Rept. No. 2a-2b; October, 1952.









Fig. 6—MUF curves for different models. a = models 5, 6, 7, 8, 9 and 10; b = models 11, 12, 13 and 14; c = models 15, 16, 17, 18, 19 and 20; d = models 21, 22, 23 and 24; e = models 25, 26, 27 and 28; f = model 30.

The formula giving the relation between verticalincidence frequency f_{\perp} and oblique incidence frequency $f \not\leq$ for a given radiation angle and a given $h_{\max'}$ [5], is

$$f \gtrless = \frac{f_{\perp}}{\sqrt{1 - \left(\frac{r_e}{r_a + h_{\max}}\sin\Phi_1\right)^2}}$$

which is in agreement with Fig. 5 and the dotted line (real height) of model 14 in Fig. 7; this can be verified by the reader.

The vertical incidence diagram demonstrates a curious fact. In Fig. 9 the traveling time as a function of the frequency for vertical incidence (in fact the ionogram) is given, in total as well as for the parts 1, 2, 3, and 4



Fig. 7—Ionograms models 14 (111) and 20 (1V). Ionograms Lindau 29-12-'54-10.00 U.T. (1) and 4-1-'55-12.30 U.T. (11).







Fig. 9—Traveling time per region as a function of frequency (vertical incidence).

of the ionosphere; part 1 being that between r_e and r_{p_1} (see Fig. 2), part 2 between r_{p_1} and r_{m_1} , part 3 between r_{m_1} and $r_{p_{22}}$ and part 4 between r_{p_2} and the maximum height reached by the ray. The vertical distances that the ray has to travel in passing parts 1 and 2 are 20 km. For parts 3 and 4 these values are 145 km for the oddnumbered models and 45 km for the even-numbered ones (see Table I). It is evident that for a rather high frequency (on the straight part of the F_2 trace of the

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Fig. 3—One-hop distances as a function of radiation angle *A*.



model 6, e.g., the frequency 16 MHz, the difference between horizontal velocity at the earth surface below 7° and above 9° is from 293,000 km/second to 290,000 km/second, that is, 1 per cent. For a total transmission distance of 8700 km or 30 m/second this means 0.3 m/second time interval between pulses. This result is consistent with experience.

The differences in horizontal velocity at earth surface are strikingly small. This velocity depends in the first place on the cosine of the radiation angle; the frequency and the model have no great influence. The well-known fact that echoes round the earth show an astonishing independency of frequency and of ionospheric situations [2], [4] is in accordance with the above-mentioned result.

The problem of multi-hop transmission poses the following considerations for us. After the first earth reflections the effective radiation angle is certainly widened considerably. Furthermore it would be necessary to take another model for the second hop-region in accordance with local conditions. In Section IV the MUF curves for one-hop transmission are given; for the second and third hop one has to take again the appropriate model.

In future practice the application of the model system might give a valuable basis for more exact path calculations in radio traffic.

IV. MUF FREQUENCIES

The formulas used to find the various desired quantities for each model include also maximum real heights for oblique and for vertical incidence. Fig. 5 gives the h_2 values for model 14 as a function of \mathcal{J}° . The MUF may be read from Fig. 5 as follows. For $\mathcal{J} = 22^\circ$ the MUF is found to lie between 18 and 20 MHz on the 200 km line (f_0F_2 maximum). Interpolation gives 18.6 MHz. For model 13, that differs from 14 only insofar as the F_2 layer is situated at 400 km instead of at 200 km, the MUF for $\mathcal{J} = 22^\circ$ appears to be 17.0 MHz (see Fig. 6), a strikingly small difference. The MUF curves of Fig. 6 have been constructed with the aid of Fig. 5.

Comparing Fig. 6 with Table 1, one will see that the influence of the strength of the E layer on the MUF is not very great. The influence of the radiation angle within a certain range is about 10 to 15 per cent. The groups of MUF curves in Fig. 6 belong to models 5 to 30 inclusive. To find the favorable hop distance it is obvious that a certain "bandwidth" of radiation angles has to be selected. The family of curves resulting from the study of the ionospheric models are consistent with the formulas and diagrams of Smith [5].

V. VERTICAL INCIDENCE- AND OBLIQUE INCIDENCE-IONOGRAMS

For all models the vertical incidence (ionogram) values have been calculated, and this makes possible a comparison of the diagrams with measured ionograms (Fig. 7). The difficulty is that the number of models (up to March 1959 67 models were available) has to be increased considerably, because, *e.g.*, *E* layers of strength 1, 2, 3, and 4 MHz do not give sufficient choice. One ought to have the possibility to choose from the series 2; 2, 2; 2, 4; 2, 6 MHz, etc. Still, it is clear that a model can be chosen which gives a good fit to the measured ionogram, In Fig. 8 recent rocket data [8], [9] for real heights have been compared with our models 22, 5, and 17 w.

TABLE	Ш
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$f = 12 \mathrm{M}$	Hz	MODEL 14 (2-8-200 km).						f = 12 M Hz	
°L	Φ_1°	h maximum	t (msec)	D km	V km/msec	t _o	t_{11}	122	
6	79.646	124.020	6.14	1795.00	292.34	1.64080	0.38214	0 49940	
8	78.381	126.799	5.32	1549.10	291.18	1 37001	0 33892	0 41298	
10	76,933	129.745	4.72	1360.50	288.24	1.16755	0.30048	0 34875	
12	75.357	132.778	4.22	1213.50	287.00	1.01320	0.26779	0.30048	
14	73.689	135.854	3.84	1096.30	285.49	0.89302	0.24042	0.26341	
16	71.956	138.945	3.56	1001.40	281.20	0.79755	0.21756	0.23430	
18	70.173	142.035	3.32	922.40	277.83	0.72032	0.19840	0.21097	
20	68.354	145.111	3.12	855,94	274.70	0.65682	0.18222	0.19192	
25	63.697	152.679	2.75	726.13	264.51	0.53938	0.15140	0.15699	
30	58.939	160.640	2.65	642.23	252.24	0.45964	0.12898	0.13348	
35	54.121	171.828	2.57	612.05	237.78	0.40270	0.11427	0.11678	

f = 12 MHz

MODEL 14 (2-8-200 km.)

f = 12 MHz

13	133	t.	2D ₀	2 <i>D</i> ₁₁	$2D_{22}$	2 <i>D</i> ₃	2D ₃₃	2 <i>D</i> ₄
$\begin{array}{c} 0.5491 \\ 0.54253 \\ 0.73779 \\ 0.53432 \\ 0.53173 \\ 0.52974 \\ 0.52819 \\ 0.52695 \\ 0.52479 \end{array}$	0.37343 0.30303	0.17657 0.35021	$\begin{array}{c} 968.70\\ 805.10\\ 682.20\\ 588.00\\ 514.00\\ 454.80\\ 406.30\\ 366.08\\ 289.93\\ 236.08\\ 195.63 \end{array}$	$\begin{array}{c} 222.23\\ 196.25\\ 173.04\\ 153.17\\ 136.41\\ 122.29\\ 110.34\\ 100.13\\ 80.24\\ 65.78\\ 54.73\\ \end{array}$	$\begin{array}{c} 288.62\\ 237.65\\ 199.59\\ 170.81\\ 148.54\\ 130.89\\ 116.61\\ 104.81\\ 82.69\\ 67.18\\ 55.59\end{array}$	$\begin{array}{c} 315.53\\ 310.25\\ 305.67\\ 301.46\\ 297.41\\ 293.36\\ 289.22\\ 284.92\\ 273.27\\ \end{array}$	185.92 142.75	87.27 163.35

trarily chosen model 14, and for frequencies 6 MHz and 10 MHz, respectively, are given. The results have been laid down in a set of curves that will be discussed hereafter.

The basic idea of this method is to find which ionospheric model, or which pair of ionospheric models, approaches the real ionosphere actually present at the given date and hour. For this purpose, it is necessary in the first place to have at one's disposal the ionogram for a station near to the reflection point, or the ionograms of two stations situated as near as possible to the points A and B. (For prediction purposes, the ionospheric quantities may be taken from the prediction charts.) For the chosen model 14, the quantities for distance D(one hop), traveling time t, maximum height h_{max} , MUF (maximum usable frequency), etc., are formed by means of the appropriate curves.

III. ONE-HOP AND MULTI-HOP DISTANCES AND TRAVELING TIMES

In Fig. 3 the one-hop distances as function of the radiation angle \bot° are given for the model 14. It can be clearly seen that the *E* layer takes a certain part of the radiated energy to abnormally great "hop" distances. For instance for 8 MHz, in the radiation angle from 9° to 11°, and for 6 MHz in the interval between 16° and 18°, this phenomenon is quite evident.

Calculations for all models and frequencies of the velocity of the radius vector at the surface of the earth, *i.e.*, the mean velocity with which the radio signal travels from .4 to B, result in the diagram of Fig. 4, giving the velocity V as a function of \mathcal{I}° . The dotted line represents the velocity $U_0 = 2D_0/2t_0$ presuming there is no ionosphere and the ray is reflected mirrorlike at the point T (see Fig. 2). The double-hatched part represents the models with F_2 maximum at 200 km, the single-hatched part gives the models with F_2 maximum at 400 km (see Table I). The lines passing from a higher to a lower curve (indicated by the frequency belonging to them) represent the transition from E-region transmission to F-region transmission. This comparatively sharp change in transmission velocity often occurs in the midst of the radiation angle, which means that a part of the energy is transported at a 1 per cent or 2 per cent lower velocity as soon as an *E* layer crossing takes place. It is obvious that for an effective radiation angle of 6° to 12° in Fig. 4 the frequency of 16 MHz is transmitted via the E layer for values below 8° and via the F_2 layer for values above 8°.

Long-distance oblique incidence pulse tests often show a tendency to produce at the receiving station two separate pulses varying in amplitude but showing a constant time interval for hours on end. This phenomenon might be explained by Fig. 4. Taking for
TABLE 1 Ionospheric Models

umber	E maxi- mum MH7	F ₂ maxi- mum	hF_2	Num- ber	E maxi- mum MH2	F_2 maxi- mum MHz	hF ₂
	MITZ	MITIZ	K I II				K III
1	4	14	400	13	2	8	400
2	4	14	200	14	2	8	200
3	2	14	400	15	4	6	400
4	2	14	200	16	4	6	200
5a	4	12	400	17	3	6	400
6a	4	12	200	18	3	6	200
7a	3	12	400	19	2	6	400
8a	3	12	200	20	2	6	200
9a	2	12	400	21	3	4	400
10a	2	12	200	22	3	4	200
5	4	10	400	23	2	4	400
6	4	10	200	24	2	4	200
7	3	10	400	25	2	- 3	400
8	3	10	200	26	2	3	200
9	2	10	400	27	1	3	400
10	2	10	200	28	1	3	200
11	4	8	400	29	1	2	400
12	4	8	200	30	1	2	200



Fig. 2-Quantities calculated.

LUDEL H	ΤA	BLE	П
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MHz		MODEL 14 (2-8-200 km).						
٥٢	Φ_1°	h maximum	t (msec)	D km	t ₀	l_1	t ₁₁	l_2
6	79.646	86.329	4.20934	1238.46	1.64080	0.46387		
8	78.381	88.156	3.66350	1072.39	1.37001	0.46174		
10	76 933	90.240	3.25628	947.25	1.16755		0.41534	0.04525
12	75.357	92.870	3,00072	866.20	1.01320		0.32043	0.16673
14	73.689	96,640	2.89606	828.17	0.89302		0.27280	0.28221
16	71.956	103.496	3.15826	892.25	0.79755		0.23938	0.54220
18	70.173	116.275	3.23058	900.97	0.72032		0.21394	
20	68.354	119.536	2.83086	780.26	0.65682		0.19375	
25	63.697	125.379	2.31598	615.42	0.53938		0.15759	
30	58.939	130.095	2.01930	511.79	0.45964		0.13366	
35	54,121	134.253	1.81952	436.40	0.40270		0.11678	
40	49.264	138,005	1.67582	375.61	0.36051		0.10437	
45	44.381	141.404	1.56850	324.30	0.32846		0.09498	
50	39.480	144.467	1.48644	279.21	0.30367		0.08774	
55	34.565	147.194	1.42284	238.35	0.28431		0.08210	

f = 6 MHz

MODEL 14 (2-8-200 km).

f = 6 MHz

122	L.	$2D_0$	$2D_1$	$2D_{11}$	2 <i>D</i> ₂	2D ₂₂	2D3	V km/msec
$\begin{array}{c} 0.41628\\ 0.30320\\ 0.20146\\ 0.15751\\ 0.13180\\ 0.11476\\ 0.10267\\ 0.09374\\ 0.08699 \end{array}$	$\begin{array}{c} 0.26475\\ 0.26166\\ 0.25956\\ 0.25884\\ 0.25848\\ 0.25848\\ 0.25827\\ 0.25814\\ 0.25807\\ 0.25807\\ 0.25807\end{array}$	$\begin{array}{c} 968.72\\ 805.13\\ 682.23\\ 587.96\\ 514.01\\ 454.76\\ 406.34\\ 366.08\\ 289.93\\ 236.08\\ 195.63\\ 163.78\\ 137.74\\ 115.75\\ 96.71\\ \end{array}$	269.74 267.26	$\begin{array}{c} 239.04\\ 183.22\\ 154.75\\ 134.54\\ 118.94\\ 106.45\\ 83.51\\ 67.68\\ 55.94\\ 46.75\\ 39.27\\ 32.98\\ 27.53 \end{array}$	25.98 95.02 159.41 302.95	$\begin{array}{c} 229.99\\ 165.54\\ 106.19\\ 79.27\\ 62.74\\ 51.08\\ 42.19\\ 35.02\\ 28.99\end{array}$	$\begin{array}{c} 145.70\\ 142.19\\ 135.88\\ 129.36\\ 122.09\\ 114.00\\ 105.10\\ 95.46\\ 85.12 \end{array}$	$\begin{array}{c} 294.21\\ 292.72\\ 290.89\\ 288.66\\ 285.96\\ 282.51\\ 278.88\\ 275.62\\ 265.72\\ 253.44\\ 239.84\\ 224.13\\ 206.75\\ 187.83\\ 167.51\\ \end{array}$

World Radio History

Ionospheric Models as an Aid for the Calculation of Ionospheric Propagation Quantities*

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Summary—This paper is a continuation of a study announced in a previous paper by the author [1]. Results are given in the form of curves, and a comparison is made with ionograms. The important influence of radiation angle on the MUF (maximum usable frequency) is shown.

A large enough number of adequate models (60 to 100) might be valuable to the radio engineer for prediction and design purposes.

I. INTRODUCTION

THE propagation path of decametric waves in the ionosphere can be calculated exactly if the electronic distribution is known. For this purpose ionospheric models consisting of E and F_2 layers of various critical frequencies have been composed. The maximum of ionization is supposed to be at 110 km for the E layers and at 400 or 200 km for the F_2 layers. The ionization curves, which start at a height of 70 km for all models, have the form shown in Fig. 1, beginning with a parabola up to half the maximum of the layer and ending with the top of an antiparabola in the E maximum and in the F_2 maximum. The parabolas are in fact third-degree curves which for the part considered have practically parabolic form (see previous articles on this subject [1], [2]).



Fig. 1—N or f_{cr} as a function of r.

From the model calculations not only interesting oblique incidence results can be obtained, but it is also possible to draw vertical incidence diagrams (ionograms) for comparison with practical ionograms obtained from ionospheric sounding stations. This comparison gives an indication of which ionospheric model represents, in the best possible way, the ionosphere that actually exists at the given moment. The method of using ionospheric models will have to be classified as an approximation to

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actual conditions, but undoubtedly it gives interesting and unexpected results of the behavior of radio propagation under normal conditions. Furthermore, it seems probable that some general rules might be derived from the results.

The formulas used for the calculations are very complicated and are given elsewhere, together with all results and curves [10].

II. The Ionospheric Models

Extensive calculations have been made for a series of 36 ionospheric models consisting of *E* layers at 110 km with critical frequencies of 2, 3, and 4 MHz and of F_2 layers at 200 or 400 km with critical frequencies of 2, 3, 4, 6, 8, 10, 12, and 14 MHz.

For each model, the parameters to calculate "hop" length, maximum height, traveling times, etc., are a series of 15 different radiation angles, including vertical incidence, and 9 frequencies.

The calculations, which have been made by electronic computers, are based on the formulas given by the author in previous papers [1], [2]. The ionization curves are third-degree equations with the earth center as origin: they have practically a parabolic form for that ionospheric part above the earth which is actually being considered.

Fig. 1 gives the electronic distribution for an ionospheric model. The *E* layer starts at 70 km parabolically up to r_{p_1} at 90 km, half-way between 70 km and the maximum at 110 km. At r_{p_1} the curve changes into a second parabolic curve and ends at the *E* maximum with different critical frequencies (see Table I, the 36 ionospheric models used for the purpose). Then, again with a third parabola, the ionization increases up to half *E* and F_2 maximum at 155 or 255 km, depending on whether the F_2 maximum height is chosen as 200 or 400 km. A fourth parabola terminates the curve in the F_2 maximum. Other models with F_2 maximum at 300 km, and also models with E_1 , F_1 and F_2 layers are in preparation.

The points r_{p_1} and r_{p_2} , where the parabolas change smoothly into antiparabolas, have been chosen at half the maximum heights. It is also possible to choose any other proportion differing from unity between $r_{p_1}-r_e$ and $r_{m_1}-r_{p_1}$, or between $r_{p_2}-r_{m_1}$ and $r_{m_2}-r_{p_2}$. Thus ionospheric models with a steep gradient or, on the contrary, with a very smooth gradient, might be introduced. The quantities which have been calculated are shown in Fig. 2; this figure needs no further explanation. In Tables II and III, extensive examples for an arbi-

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lector voltage, $V_{cm}/V_{\alpha M\beta}$, vs normalized power dissipation $s\phi P$ for an RCA developmental *p*-*n*-*p* germanium transistor. The "calculated" curve is based on the value of n = 4.0 which was determined from measured values of (M-1)/M vs V_c for this transistor (see Fig. 13).



Fig. 11—Measured values of alpha multiplication breakdown voltage $V_{\alpha M\beta}$ vs circuit stability factor S_f for an RCA developmental $n \cdot p \cdot n$ silicon transistor (see Fig. 9).



Fig. 12—Comparison between measured and predicted values of normalized maximum stable collector voltage V_{cm}/V_{aMB} vs normalized power dissipation $s\phi P$ for an RCA developmental *p*-*n*-*p* germanium transistor. (The "calculated" curve is based on n = 4.0 as determined in Fig. 13.)





yield the collector junction avalanche breakdown voltage V_b which corresponds to an infinite value of M.

The measurement of (M-1)/M was accomplished by measuring the ratio of the small signal base current to the collector currents as shown in Fig. 8. The base current may be found from (4) and (5) as follows:

$$I_{b} = - (I_{c} + I_{e}) \equiv I_{c} \left[\frac{1 - \alpha_{N}M}{\alpha_{N}M} \right] + \frac{I_{co}}{\alpha_{N}} \cdot \quad (43)$$



Fig. 8—A circuit for the measurement of n.

If the resistances Rc and R_B are small, then the ratio of the small signal base current to collector current is given by

$$\left|\frac{\partial I_b}{\partial I_c}\right| = \frac{R_c}{R_B} \left|\frac{V_b}{V_c}\right| = \left|\frac{\alpha_N M - 1}{\alpha_N M}\right|,\tag{44}$$

which may be measured as a function of the dc collector junction voltage V_c . After measuring α_N (at some low voltage) the value of (M-1)/M may then be computed.

Employing the above technique, the variation of (M-1)/M was measured as a function of V_c with the emitter current held constant, using an RCA developmental *n-p-n* silicon transistor. The results which are shown plotted in Fig. 9, indicate a straight line having a slope of *n* equal to 4.8 and V_b equal to 195 volts.

B. Measurement of $V_{\alpha M\beta}$ vs S_F —Verification of (27)

Eq. (27), which expresses the multiplication breakdown voltage $V_{\alpha,M\beta}$, as a function of the circuit stability factor S_F was verified experimentally using the circuit shown in Fig. 10.

The circuit stability factor S_F of the circuit under test was varied by changing the ratio of R_E to R_B . The multiplication breakdown voltage $V_{\alpha M\beta}$ of the circuit under test may be defined as the collector-to-base voltage necessary to produce an ac null voltage across resistance R. At this collector voltage, the current through R no longer controls the collector current. At higher collector junction voltages, the incremental current gain of the circuit under test is greater than unity and would result in unstable operation if the emitter current were not limited by R_E . Hence this voltage represents the



Fig. 9—Measured values of (M-1)/M vs V_{cb} for an RCA developmental *n-p-n* silicon transistor (see Fig. 8).



Fig. 10—Circuit for the measurement of $V_{\alpha M\beta}$ of the transistor circuit shown in the dotted box.

alpha multiplication-breakdown voltage $V_{\alpha M\beta}$ for the circuit under test. The over-all circuit should, of course, have a low value of stability factor to insure over-all stability.

Fig. 11 shows a plot of measured values of $V_{\alpha MB}$ as a function of S_F . The stability factor was calculated for the circuit under test using (29). When plotted to a loglog scale, (27) should be expressible as a straight line having a slope of -1/n. Extrapolation of this line to the point where S_F equals unity (open emitter) should likewise yield the avalanche breakdown voltage of the collector junction, V_b . As is evident from the experimental curve, a straight line is obtained corresponding to the same value of n determined above (n=4.8) and by extrapolation indicates a collector junction breakdown voltage, $V_b = 185$ volts. The value of V_b obtained using an electronic curve tracer was 190 volts. It should be mentioned that excellent experimental agreement was also obtained for p-n-p germanium transistors.

C. Measurement of the Maximum Stable Collector Voltage Derating Characteristic (42)

Fig. 12 shows a comparison between measured and predicted values of normalized maximum stable col-

It is also assumed that the transistor is forward biased so that (26) is satisfied and is operating under the following conditions:

Heat-sink temperature, T _s	45°C
Power dissipation, P	13.5 watts
Emitter-circuit resistance, R _e	1 ohm
Base-circuit resistance, R_h, \ldots, \ldots	10 ohms.

The feedback factor from collector to base β_{be} is found by considering the "black box" equivalent circuit shown in Fig. 7. By inspection,

$$\beta_{bc} = I_B / I_C = \frac{R_E}{R_E + R_B + r_{bb'}} = \frac{1}{1 + 10 + 39} = 0.02.$$

From (29), the stability factor is



Fig. 7—Current transfer ratio of the "black box" of of the circuit shown in Fig. 4.

The breakdown voltage in the absence of thermal heating is given by (34):

$$V_{\alpha M\beta} = V_{\alpha M} (1 + \alpha_{cb} \beta_{bc})^{1/n}$$

= 40(1 + 75 × 0.02)^{1/3} = 54 volts.

The value of V_s is given by (39):

$$V_s = 1/[s\phi S_f I_{so} e^{\phi(T_s - T_o)}]$$

= 1/[(1.7)(0.1)(30)(50 × 10^{-6})e^{0.1(45-25)}]

= 530 volts.

Therefore,

$$V_s/V_{\alpha M\beta} = 530/54 \cong 10.$$

The power dissipated is 13.5 watts, hence

$$s\phi P = (1.7)(0.1)(13.5) = 2.3.$$

Once the values of $V_s/V_{\alpha,M\beta}$ and $s\phi P$ have been computed, the maximum stable collector voltage V_{cm} is found directly from the graph of Fig. 6. In this case $V_{cm}/V_{\alpha,M\beta}$ is found to be 0.68, or $V_{cm}=37$ volts.

When the maximum stable collector voltage is computed neglecting thermal effects, a value

$$V_{cm}]_{s=0} = 54$$
 volts

is obtained which is the alpha multiplication breakdown voltage $V_{\alpha,M\beta}$. When the maximum stable collector voltage is computed neglecting alpha multiplication, a value

$$V_{cm}]_{M=1} = V_s e^{-s\phi P} = 530 e^{-(1.7)(0.1)(13.5)} = 53$$
 volts

is obtained. A collector voltage equal to say 45 volts appears to be well within both stable limits considering the effects separately. This example, however, demonstrates a considerable degradation of the expected maximum stable collector voltage caused by the interaction of both alpha multiplication and thermal effects simultaneously present. In spite of the fact that a collector voltage of 45 volts exceeds neither the multiplication breakdown voltage (54 volts) nor the thermal runaway voltage (53 volts), it does exceed the combined derated voltage limit of 37 volts obtained from the graph of Fig. 6 and would result in instability and eventual destruction of the transistor.

Experiments

Three experiments were performed in order to verify the predictions of the above theory. The first of these was designed to verify (6) and thereby measure the value of the empirical constant n. Substitution of this value of n into (27) defines the relationship between the alpha multiplication-breakdown voltage $V_{\alpha M\beta}$ and the circuit stability factor S_F . The second experiment was therefore designed to verify whether (27), with the value of *n* determined above, actually predicts the measured values $V_{\alpha M\beta}$ vs S_F . The details of the experimental procedure employed for these measurements and the results obtained are described below. Finally, the maximum collector voltage derating curve of Fig. 6 was measured and compared with theory for an RCA developmental germanium *p-n-p* junction transistor. These results are plotted in Fig. 13.

A. Measurement of n—Verification of (6)

The value of *n* can be determined experimentally by plotting the measured quantity (M-1)/M vs collector junction voltage V_e to a log-log scale. Eq. (6) indicates that this function can be represented by a straight line of slope *n*. Furthermore, extrapolation of this line to the point where the value of (M-1)/M equals unity should

where T_s is the temperature of the heat sink. When (35)-(37) are combined, the maximum thermally stable collector voltage is found to be given by

$$V_{cm} = V_s e^{-s\phi P}, \qquad (38)$$

where

$$V_s = \frac{1}{s\phi S_f I_{so} e^{\phi (T_s - T_o)}}$$
 (39)

It is evident that in the absence of multiplication effects, the maximum collector voltage that can be safely applied to the transistor decreases exponentially with power dissipation and is inversely proportional to the thermal resistance, stability factor, and saturation current I_{ee} measured at sink temperature.

DETERMINATION OF MAXIMUM STABLE COLLECTOR VOLTAGE WHEN MULTIPLICATION AND THERMAL EFFECTS ARE CONSIDERED

When both heating and multiplication effects are present, the maximum collector voltage may again be determined from the stability criterion given by (20) with M greater than unity and s different from zero. Under normal operating conditions the emitter junction is forward biased so as to satisfy (26); the stability criterion now becomes

$$1 - \alpha_N M \beta_{ec} - s \phi V_{cm} M I_{co} = 0, \qquad (40)$$

which can be rearranged as

$$V_{cm} = \frac{1 - \left[\frac{1 - 1/M}{1 - \beta_{ec}\alpha_N}\right]}{\left[\frac{s\phi I_{co}}{1 - \beta_{ec}\alpha_N}\right]}$$
(41)

When (41), is combined with (6), (27), (28) and (39),

$$V_{cm} = \left[1 - (V_{cm}/V_{\alpha M\beta})^n\right] V_s e^{-s\varphi P}, \qquad (42)$$

where $V_{\alpha M\beta}$ and V_{δ} are defined by (27) and (39), respectively.

The stability criterion given by (42) is represented graphically in Fig. 6 for n = 3. With the aid of this graph, it is possible to determine the maximum stable collector voltage when both multiplication and thermal effects are significant. This graph expresses the ratio $V_{cm}/V_{\alpha M\beta}$ in terms of normalized power dissipation, $s\phi P$, with a running parameter $V_s/V_{\alpha M\beta}$. When the thermal runaway voltage $V_s e^{-s\phi P}$ is large relative to the value of the multiplication breakdown voltage $V_{\alpha M\beta}$ the maximum stable collector voltage is limited by the multiplication



Fig. 6—Derating curves for determining maximum stable collector voltage for n = 3.

mechanism, and the value of V_{cm} approaches that given by (27). As the thermal runaway voltage $V_s e^{-s\phi P}$ is made small relative to the multiplication breakdown $V_{\alpha,M\beta}$ the maximum stable collector voltage becomes thermally limited and the value of V_{cm} approaches that given by (38). V_{cm} now tends to decrease exponentially with power dissipation. This decrease is caused by I_{co} increasing with junction temperature. If the thermal resistance is decreased, the rise in junction temperature required to dissipate the generated heat is reduced and, therefore, the value of the maximum stable collector voltage is increased.

ILLUSTRATIVE EXAMPLE

The following example illustrates the calculation of the maximum stable collector voltage that can be applied to a p-n-p transistor based on a knowledge of the circuit configuration, mode of emitter bias, rate of heat dissipation, and ambient temperature.

The transistor is the circuit shown in Fig. 4 is assumed to have the following characteristics:

Thermal resistance, s	1.7 °C/watt
Empirical exponent, <i>n</i>	3.0
Collector-to-base alpha, α_{cb}	75
Saturation current, <i>I</i> _{so}	50 µamps
measured at temperature, T_o	25°C
Temperature coefficient of I_{so} , ϕ	0.1 per °C
Breakdown voltage for $\alpha_{cb}R_e/R_b \ll 1$,	
$s=0; V_{\alpha M}$	40 volts
Base-lead resistance, rbb'	39 ohms.

eakdown voltage again

Under these conditions, the breakdown voltage becomes independent of the emitter-junction bias (V_e or R_{eb}), and depends almost entirely on the amount of feedback present in the circuit. Breakdown now occurs when $\alpha_N M\beta_{ee} = 1$. The maximum stable collector voltage is now equal to the multiplication-breakdown voltage, $V_{\alpha M\beta}$. Combining (23) and (26) yields

$$V_{\alpha,M\beta} = V_b / (S_f)^{1/n} = V_{cm} \Big]_{\substack{s=0\\\alpha N,M\beta ec=1,}}$$
(27)

where S_f is defined as the stability factor,⁶

$$S_f = \frac{1}{1 - \alpha_N \beta_{ee}}$$
 (28)

The value of the stability factor S_f is of great importance, therefore, in determining the multiplication breakdown voltage. For numerical calculations, it is more convenient to express S_f in the form

$$S_f = \frac{1 + \alpha_{cb}}{1 + \alpha_{cb}\beta_{bc}},\tag{29}$$

where

$$\alpha_{cb} \equiv \frac{\alpha_N}{1 - \alpha_N} \tag{30}$$

is the low-voltage transistor current gain from base to collector, and

$$\beta_{bc} \equiv 1 - \beta_{ec} \tag{31}$$

is the current-feedback factor from collector to base in the circuit, as defined previously.

It is evident that the stability factor approaches its maximum value as the collector-to-base feedback factor β_{bc} is made small relative to $1/\alpha_{cb}$. The breakdown voltage for this circuit condition approaches $V_{\alpha,M}$ where

$$V_{\alpha M} = \frac{V_b}{(1 + \alpha_{cb})^{1/n}} \equiv V_{cm} \bigg]_{\substack{\alpha_N M = 1 \\ s = 0 \\ \beta_{cc} = 1}}.$$
 (32)

This breakdown voltage occurs when $\alpha_N M$ equals unity, which represents the least stable circuit condition. V_c need only be sufficient to increase M from unity to $1/\alpha_N$ to initiate breakdown where α_N ranges from about 0.95 to 0.99 in junction transistors.

As the collector-to-base feedback factor, β_{bc} is made large relative to $1/\alpha_{cb}$, the value of S_f approaches unity and the breakdown voltage again approaches that of the collector junction V_b and corresponds to an infinite value of M. Thus, in the absence of thermal heating, the breakdown voltage $V_{\alpha,M\beta}$ can approach its maximum value V_b as the circuit stability factor approaches unity or when the emitter junction is reversed biased or "floating." In the intermediate case, where the emitter junction is not biased sufficiently in either polarity, conditions (24), (25) and (26) do not apply, and V_{cm} has a value which must be calculated directly from (23). Thus, by changing the circuit-stability factor or the mode of bias of the emitter junction, the multiplication breakdown can be controlled from V_b to $V_b/(1+\alpha_{cb})^{1/n}$, a ratio as large as four to one.

As previously explained, it may not be possible to measure V_b directly in practice because of surfacebreakdown mechanisms which may occur at voltages lower than V_b . It is more desirable, therefore, to express $V_{\alpha,M\beta}$ in terms of $V_{\alpha,M}$ rather than V_b . Combining (27) and (32) gives

$$V_{\alpha M\beta} = V_{\alpha M} \left[\frac{1 + \alpha_{cb}}{S_f} \right]^{1/n}.$$
 (33)

When this equation is combined with (29), a more convenient form of (27) results:

$$V_{\alpha M\beta} = V_{\alpha M} (1 + \alpha_{cb} \beta_{bc})^{1/n}. \tag{34}$$

MAXIMUM STABLE COLLECTOR VOLTAGE WHEN ONLY THERMAL EFFECTS ARE CONSIDERED

When avalanche-multiplication effects within the collector junction are negligible (at low collector voltages), the maximum collector voltage is limited by the onset of thermal runaway. The conditions for thermal stability of the transistor at low voltages can be investigated by using the stability criterion given by (20) with M equal to unity. Under normal operating conditions, the emitter junction is forward biased so as to satisfy (26); (20) reduces to

$$\frac{1}{S_f} - s\phi V_c I_{co} = 0. \tag{35}$$

The collector saturation current I_{co} has approximately the same temperature coefficient as I_{cs} and I_{cs} and increases at a rate of about 10 per cent per degree Centigrade so that at junction temperature T_{cs}

$$I_{co} = I_{so} e^{\phi (T - T_o)}, \tag{36}$$

where I_{so} is the value of I_{co} at temperature T_o . Under equilibrium conditions, the junction temperature rises above the sink temperature by an amount equal to the power dissipated times the thermal resistance,

$$T - T_s = sP, \tag{37}$$

⁶ R. F. Shea, "Transistor operation: stabilization of operating points," PROC. IRE, vol. 40, pp. 1435-1437; November, 1952.



Fig. 2—Resistance R_{cb} of the "black box."



Fig. 3-Current transfer ratio of the "black box."



Fig. 4-DC circuit of a typical transistor stage.



Fig. 5—Signal flow graph, $\Lambda \equiv q/kT$.

The peak reverse-bias collector voltage is reached when $dI_c/dV_c = \infty$ or when the denominator of (19) approaches zero,

$$1 - s\phi V_c I_c + (q/kT) R_{eb} I_{es} e^{qV_c/kT}$$

$$\cdot \left[1 - \alpha_N M \beta_{ec} - s\phi V_c M I_{co} \right] = 0. \quad (20)$$

Eq. (20) is the stability criterion from which the maximum stable collector voltage can be derived under various operating conditions.

DETERMINATION OF THE MAXIMUM STABLE COLLECTOR Voltage in the Absence of Thermal Heating Effects

When the thermal resistance of the transistor is small enough, the maximum reverse-bias voltage can be derived from (20) by setting s equal to zero which gives

$$1 + (q/kT) R_{eb} I_{es} e^{qV_e/kT} [1 - \beta_{ee} \alpha_N M] = 0.$$
 (21)

Eq. (21) is then solved for *M* to determine the amount of multiplication required to make the circuit unstable,

$$M = \frac{1}{\alpha_N \beta_{\iota e}} \left[\frac{1 + (q/kT) R_{eb} I_{\iota s} e^{qV_{\iota/kT}}}{(q/kT) R_{eb} I_{\iota s} e^{qV_{\iota/kT}}} \right].$$
(22)

The collector-to-base voltage corresponding to M is found by solving (6) and (22) for U_c ,

$$V_{cm} = V_b \left[\frac{1 + \left[(q/kT) R_{cb} I_{sb} e^{qV_{s}/kT} \right] \left[1 - \alpha_N \beta_{cc} \right]}{1 + \left[(q/kT) R_{cb} I_{cs} e^{qV_{E}/kT} \right]} \right]^{1/n}, (23)$$

where V_{em} is the value of V_e above which the circuit becomes unstable.

There are two sufficient conditions for which the circuit can be stabilized for all values of V_c up to V_b . These conditions are that

$$(q/kT)R_{eb}I_{es}e^{qV_c/kT} \ll 1, \qquad (24)$$

or that

$$\alpha_N \beta_{cc} \ll 1. \tag{25}$$

When either of these conditions is satisfied, the collector voltage must approach the avalanche-breakdown voltage of the collector junction in order to produce breakdown. Condition (24) is satisfied if a reverse-bias voltage of a few tenths of a volt is maintained across the emitter junction, or if the dc circuit resistance between the intrinsic base-to-emitter terminals $R_{,b}$ is made sufficiently small so that the intrinsic emitter-junction voltage is held constant. The condition given by (25) requires that the current-feedback factor β_{ee} be small, *i.e.*, that the ratio of series base resistance to emitter resistance be small. Either of these conditions tends to stabilize the emitter current and thereby yield a breakdown voltage approaching V_b .

When the circuit conditions are such that the emitter junction is forward biased through a dc source resistance R_{cb} of sufficient magnitude for the voltage drop across base and emitter circuit resistances to exceed a few tenths of a volt, then

$$(q/kT)R_{eb}I_{es}e^{qV_e/kT}\left[1-\alpha_N\beta_{ec}\right]\gg 1.$$
(26)

Taking the total differentials of (4) and (5) yields where

$$dI_c = \frac{I_c}{M} dM + \phi I_c dT - \frac{q}{kT} \alpha_N M I_{es}(e^{qV_c/kT}) dV_e \quad (7)$$

and

$$dI_{e} = \phi I_{e} dT + \frac{q}{kT} I_{es}(e^{qV_{e}/kT}) dV_{e}, \qquad (8)$$

where

$$dM = \frac{nM(M-1)}{V_c} dV_c.$$

If it is assumed that under dc conditions the junction is allowed to attain thermal equilibrium at each point of the characteristic, the heat generated is equal to the heat dissipated from the junction, and Newton's law of cooling (dT/dP = s) is valid. Then,

$$dT \cong s V_c dI_c, \tag{9}$$

where *s* is the thermal resistance in °C/watt.

Eqs. (7)–(9) describe the effect on the intrinsic transistor of differential changes in the variables V_c , I_c , V_e and T. Any extrinsic effects, such as leakage resistance or base spreading resistance, may be included in the external circuit of the transistor. It is assumed that the external circuit of the transistor contains only constant resistances and voltage sources. When, as shown in Fig. 1, the circuit external to the intrinsic transistor is considered as contained in a "black box" having the same terminal currents and voltages as the transistor,

$$dI_e = \frac{\partial I_e}{\partial I_c} \bigg|_{V_e} dI_c + \frac{\partial I_e}{\partial V_e} \bigg|_{I_c} dV_e$$
(10)

$$dI_e = -\beta_{ec}dI_c - (1/R_{eb})dV_e$$

 $\boldsymbol{\beta}_{\epsilon}$

$$_{e} = -\frac{\partial I_{e}}{\partial I_{e}}\Big|_{V_{e}}$$
(12)

$$1/R_{eb} = -\frac{\partial I_e}{\partial V_e}\Big|_{I_c \text{ constant}}.$$
 (13)

The circuit parameters R_{cb} and β_{ec} describe the properties of the linear network contained in the "black box." It is evident that R_{cb} and β_{ec} are completely independent of the properties of the intrinsic transistor. R_{cb} is the dc resistance of the "black box" measured between the emitter and base terminals with the collector terminal "floating," and with the voltage sources replaced by thei internal resistances, as shown in Fig. 2. β_{ec} is the dc current transfer ratio of the "black box" from the collector to emitter terminals with the voltage sources again replaced by their internal resistances and with the emitter and base terminals short-circuited as shown in Fig. 3. Another useful circuit parameter is the current transfer ratio from collector to base, β_{bc} , which is measured in a manner similar to β_{ec} .

The circuit shown in Fig. 4 can be used as an example. It is evident on inspection that

$$R_{eb} = R_e + R_b + r_{bb'}, (14)$$

$$\beta_{ec} = \frac{R_b + r_{bb'}}{R_b + R_c + r_{bb'}}.$$
 (15)

$$\beta_{bc} = \frac{R_e}{R_b + R_e + r_{bb'}}$$
 (16)

In general,

(11)

$$\beta_{bc} + \beta_{ec} \equiv 1. \tag{17}$$

Eqs. (7), (8), (9) and (11) may be solved for dI_c/dV_c , thus eliminating the variables dV_c , dI_t and dT. The solution can be obtained by inspection with the aid of the signal-flow graph representation shown in Fig. 5:

$$\frac{dI_c}{dV_c} = \frac{\left[n(M-1)I_c/V_c\right]\left[1 + (q/kT)R_{eb}I_{cs}e^{qV_c/kT}\right]}{1 - s\phi V_c I_c + (q/kT)R_{eb}I_{es}e^{qV_c/kT}\left[1 - \beta_{ec}\alpha_N M - s\phi V_c(I_c + \alpha_N M Ie)\right]}$$
(18)

The term $(I_c + \alpha_N M I_e)$ is equal to $M I_{co}$ where I_{co} is the reverse-biased collector-current of the transistor at low voltages (M = 1), measured with the emitter floating. By substitution, (18) becomes

$$\frac{dI_c}{dV_c} = \frac{\left[n(M-1)I_c/V_c\right]\left[1 + (q/kT)R_{eb}I_{es}e^{qV_c/kT}\right]}{1 - s\phi V_c I_c + (q/kT)R_{eb}I_{es}e^{qV_c/kT}\left[1 - \beta_{ec}\alpha_N M - s\phi V_c M I_{co}\right]}$$
(19)

or

generalized junction transistor, such as that shown in Fig. 1, the equations are

$$I_{c} = -\alpha_{N}I_{es}(e^{qV_{e}/kT} - 1) + I_{cs}(e^{qV_{c}/kT} - 1)$$
(1)

and

$$V_{e} = + I_{ss}(e^{qV_{e}/kT} - 1) - \alpha_{I}I_{cs}(e^{qV_{c}/kT} - 1), \quad (2)$$

where

 I_c = collector junction current;

- I_e = emitter junction current;
- α_N = normal alpha, defined as the ratio of collector current to emitter current when the intrinsic collector junction is held to zero volts;
- α_l = inverted alpha, defined as the ratio of emitter current to collector current when the intrinsic emitter junction is held to zero volts;
- $I_{\epsilon s}$ = emitter junction diode saturation current;
- I_{cs} = collector junction diode saturation current;
- q/kT = where q is the electron charge, k is Boltzmann's constant, and T is absolute temperature (numerically, q/kT = 38.5 per volt at 25°C); V_e = intrinsic collector junction voltage; V_e = intrinsic emitter junction voltage.

When the collector junction is reversed-biased, qV_c/kT is less than zero, and the term $e^{qV_c/kT}$ becomes quite small. In the range of collector voltages from a few tenths of a volt to several volts, (1) and (2) are essentially independent of V_c , so that

$$I_{c} = -\alpha_{N}I_{es}(e^{qV_{e}/kT} - 1) - I_{cs}$$
(3)

and

$$I_{e} = I_{es}(e^{qV_{e}/kT} - 1) + \alpha_{I}I_{cs}.$$
 (4)

As the reverse-bias voltage V_c is increased still further, the collector current increases because of multiplication of the charge carriers flowing across the collector junction. To account for this effect, (3) must be multiplied by a factor M which is defined as the ratio of actual current to the current which would flow if multiplication were not present. The collector current is then

$$I_c = - \alpha_N M I_{es} (e^{qV_c/kT} - 1) - M I_{cs}. \qquad (5)$$

The multiplication factor M increases with V_c in accordance with the empirical formula^{1,2}

$$M = \frac{1}{1 - (V_c/V_b)^n},$$
 (6)

where U_b is the ultimate avalanche-breakdown voltage of the collector junction and depends upon the impurity density in the base material, and *n* is a constant equal to about 3 for germanium *p*-*n*-*p* junction transistors and ranging from 4.6 to 6.6 for germanium *n*-*p*-*n* junction transistors. The observed breakdown voltage of the collector-junction diode is not always equal to V_b . Breakdown mechanisms other than the avalanche type exist



Fig. 1—Generalized transistor circuit showing the intrinsic transistor and its connection to the "black box."

(highly nonlinear surface conduction, for example), and breakdown may occur at a lower voltage. For low voltages, M is close to unity because very little carrier multiplication occurs. As the voltage increases, more carrier multiplication takes place, and the multiplication factor increases until V_c equals V_b , at which point M becomes infinite.

In practice, M can never become infinite as this would imply an infinite rate of heat generation in the transistor. Before this point is reached, the transistor becomes thermally unstable and experiences thermal runaway. Because the actual breakdown is inherently thermal, the effect of M on the thermal stability of the transistor in the prebreakdown region is of great importance in the determination of the maximum stable collector voltage.

TEMPERATURE CHARACTERISTICS

The dc characteristics of the transistor represented by (4) and (5) are temperature-sensitive, primarily because the diode-saturation currents I_{es} and I_{cs} vary with temperature. These currents have temperature coefficients equal to $e^{\phi dT}$ where

$$\phi \cong qE_{g}/kT^{2} \cong 0.1$$
 per degree C at 25° C.

 E_g is the forbidden energy gap potential; for germanium, $E_g = 0.7$ volt. At room temperature, I_{es} and I_{rs} increase approximately 10 per cent per degree Centigrade.⁵ For simplicity, these two quantities are assumed to be the only temperature-sensitive variables of the transistor.

The maximum reverse voltage is reached when a differential increment in V_e causes an infinite increase in collector current, *i.e.*, when

$$\frac{dI_c}{dV_c} = \infty,$$

where V_c is considered to be the independent variable.

⁶ H. C. Lin and A. A. Barco, "Temperature effects in circuits using junction transistors," *Transistors I*, RCA Labs., Princeton, N. J., pp. 369-402; March, 1956.

Maximum Stable Collector Voltage for Junction Transistors*

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Summary—A study is made of the conditions for stability of junction transistors operating in the pre-avalanche breakdown region in order to establish on quantitative grounds the maximum collector voltage that can be applied to the transistor for stable operation. A method is shown by which this peak voltage can be calculated from a knowledge of the circuit configuration, mode of emitter bias, rate of heat generation, and ambient temperature.

INTRODUCTION

PERATION of junction transistors at high voltages has become more practical recently as a result of improved techniques which make possible devices having lower leakage currents, lower thermal resistance, and, therefore, better temperature stability. There is, however, an inherent limit to the collector junction voltage that is allowable for stable operation of junction transistors. When the collector voltage reaches a high value, a current-multiplication effect occurs within the collector junction.¹ This phenomenon may cause excessive heating, instability, and eventual breakdown.

The mechanism of avalanche breakdown involves a charge-carrier multiplication effect which results when holes and electrons attain sufficient energy because of the high electric field present, to interact with valence electrons and produce electron-hole pairs.² This process is cumulative and eventually results in a breakdown of the junction similar to the ionization in gas discharge tubes. The ionization rate, which increases with the electric field strength, is entirely negligible for small reverse biases, but increases rapidly at higher voltages until a critical voltage is reached at which the junction current avalanches uncontrollably. The avalanche-breakdown voltage, therefore, defines the maximum reverse bias voltage that can be applied to the collector junction under any conditions.

In the prebreakdown region, the current-multiplication effect is not sufficient to produce an avalanche breakdown of the collector junction but may have a profound effect on the stability of the transistor. As the reverse-bias collector-junction voltage is increased, the emitter to collector dc alpha increases because of the multiplication effect and may, in general,

† Semiconductor and Materials Division, RCA, Somerville, N. J. ¹ M. C. Kidd, W. Hasenberg, and W. M. Webster, "Delayed collector conduction—a new effect of junction transistors," *RCA Rev.*, be less than, equal to, or greater than unity depending on the magnitude of the collector-junction voltage.³ When the transistor is operated so that its alpha exceeds unity, the circuit may be unstable and produce a premature breakdown of the transistor at voltages that are considerably less than that necessary to cause an avalanche breakdown of the collector junction itself. Because the application of dc negative feedback can be used to degenerate the effective gain of the transistor, it is possible to design the circuit so that stable operation is maintained with values of alpha greatly exceeding unity. As the collector voltage approaches the avalanche breakdown voltage of the collector junction, however, the alpha becomes infinite and no amount of feedback can stabilize the collector current. The maximum reverse-bias voltage that can safely be applied to the collector junction in a given application is not fixed, therefore, but is controlled by circuit design.

The thermal stability of the transistor is also adversely affected by the multiplication phenomenon. Because the multiplication phenomenon increases the collector-saturation current as well as the current gain of the transistor, an increase in collector voltage may cause a premature thermal breakdown which would not ordinarily occur. In such cases, the breakdown is caused by thermal instability in the presence of the multiplication mechanism. The relative importance of the two effects is dependent upon such parameters as the collectorjunction avalanche-breakdown voltage, dc alpha, the amount of negative feedback present in the circuit, thermal resistance, saturation current, ambient temperature, and rate of heat generation.

An analysis of the requirements for operational stability of the transistor, therefore, must include the effects both of thermal heating and of the multiplication phenomenon. In this paper, the collector-breakdown voltage is defined as the maximum collector-to-base voltage at which the transistor remains stable, regardless of whether this maximum voltage is limited by thermal runaway or by multiplication breakdown.

DERIVATION OF STABILITY CRITERION

The intrinsic dc characteristics of junction transistors in the absence of avalanche-multiplication effects and leakage currents have been described elsewhere.⁴ For the

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<sup>vol. 16, pp. 16–33; March, 1955.
² S. L. Miller, "Avalanche breakdown in germanium,"</sup> *Phys. Rev.*, vol. 99, pp. 1234–1241; August, 1955.

³ H. Schenkel and H. Statz, "Junction transistors with alpha greater than unity," PROC. IRE, vol. 44, pp. 360-371; March, 1956. ⁴ J. J. Ebers and J. L. Moll, "Large signal behavior of junction transistors," PROC. IRE, vol. 42, pp. 1761-1772; December, 1954.

Day

At first sight it would appear that the experimental results for day are in good agreement with simple theory using a constant height, since the quantity \sqrt{f}/α does not change appreciably in Fig. 1 between 300 and 1000 cps. However, consideration indicates that because $h\sqrt{\omega}_r$ is practically constant over an appreciable range in height by day (Fig. 3), the experimental results are not entirely inconsistent with the idea of a height changing with frequency as postulated in the preceding section.

The argument is as follows. First of all, to clarify matters, the values of $h\sqrt{\omega_r}$ corresponding to the observed attenuation coefficients at 100, 300, and 1000 cps, are plotted on curve B of Fig. 3. Now the graphs of Fig. 2 giving the variation of N and γ with height are by no means precise; for example, at 70 km, values of γ of between 1×10^7 and 3×10^7 collisions per second have been quoted in the past. Uncertainties of this order imply that the curves of Fig. 3 could well be displaced at least within the limits indicated by the dotted lines. Such a displacement would have relatively little influence upon the interpretation of the night results since the graph A on Fig. 3 changes little in slope. By day, however, the situation is entirely different. In particular, only a slight displacement is needed to bring the experimentally determined values of $h\sqrt{\omega_r}$ for 300 and 1000 cps on to the steepest part of curve B. It is tempting to suggest that this is indeed the case since under these circumstances a change in H of roughly 2.5 km between 300 and 1000 cps as indicated by the night results, would correspond by day to a variation of only about 5 per cent in $h\sqrt{\omega_r}$, and therefore a change also of the same order in the experimentally determined $\sqrt{f/\alpha}$. The scatter and uncertainties in experimental results such as those of Chapman and Macario are probably sufficient to obscure variations of this magnitude. Thus it is possible, without too great difficulty, to reconcile the concept of a height varying with frequency with the experimental results both by day and by night. It must be admitted, however, that if the apparent minimum of about 600 cps for $\sqrt{f/\alpha}$ by day (see Fig. 1) is a real effect, then the derived H is not monotonic with frequency, and the reinterpretation may be said to fail for the day results.

Discussion

It has been pointed out how there are difficulties in the interpretation on the simple waveguide theory of the experimental results at night for the attenuation of ELF radio waves during propagation. The difficulties are removed by the physically plausible postulation that the height of the guide increases as the frequency decreases. This concept has also been shown to be not entirely inconsistent with the daytime results.

The two layer model introduced by Wait has already been mentioned. This model referred primarily to propagation in the VLF range, but Wait has also considered a second extension to the mode theory for ELF propagation, being concerned with the indication from the experimental work of Holzer, Deal, and Ruttenberg that there is a minimum of attenuation at around 100 cps. Incidentally, the first diagram in the paper by Chapman and Macario also suggests that the daytime attenuation is, if anything, less at 125 cps than at either 100 or 160 cps, although Chapman and Macario in their derived attenuation coefficients give a monotonic increase from 100 to 1000 cps. In his second model, Wait retains the concept of a waveguide with a sharply bounded edge at the lower ionosphere but adds an exponential variation of N/γ decreasing upwards within the ionosphere. Strictly speaking, this picture is untrue at least up to the maximum of electron density in the F region; effectively, however, it may be legitimate over distances comparable with a wavelength (3000 km for 100 cps), although this is not immediately evident bearing in mind the comparatively slow decrease in N above the F2 maximum indicated by Sputnik observations and by whistler results for farther from the earth, together with the uncertainties relating to the values for γ . The tapered exponential model of Wait can account for a minimum of attenuation at a frequency of the order of 100 cps and in this respect has the advantage over the approach advanced in the present paper. In general, however, the latter seems to be more effective in explaining the discrepancies, particularly at night, between Chapman and Macario's results and those anticipated on the simple theory.

It is, however, perhaps academic to pursue the development of propagation theories at these low frequencies while the experimental observations remain so scanty. Further and more detailed results are urgently needed and these could be immediately obtained in the manner of Chapman and Macario by work on atmospherics. Multistation observations would be even more valuable.

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Fig. 1—Relation to frequency of Chapman and Macario's experimental values for the attenuation coefficient α .

THE REINTERPRETATION OF THE RESULTS

Night

The quantity $h\sqrt{\omega_r}$ may be estimated directly from ionospheric data for the variation of N and γ with height. The exact form of this variation is not known but the situation has been summarized by Waynick⁵ who gives several references to the best available information. A combination of this data is depicted in Fig. 2, while Fig. 3 represents the variation of $h\sqrt{\omega_r}$ with height, for both day and night conditions, deduced from the curves of Fig. 2. It can be seen from Fig. 3 that $h\sqrt{\omega_r}$ increases monotonically with increasing height, but that the rate of increase is very slight by day for the range of height 71 to 77 km.

It is interesting to accept Chapman and Macario's experimental figures for \sqrt{f}/α from Fig. 1, and then use (1) to obtain the corresponding values of $h\sqrt{\omega_r}$ at various frequencies. In turn these values may be employed to define a series of heights since $h\sqrt{\omega_r}$ is monotonic with height. These are indicated, for frequencies at intervals of 100 cps, by the horizontal lines on the night curve Λ of Fig. 3; for these night results the variation within the range 100 to 1000 cps is represented quite well by the empirical relation

$$H = (89 - 0.17\sqrt{f}), \tag{2}$$

or with rather less accuracy by

$$II = \left(81.0 + \frac{85}{\sqrt{f}}\right). \tag{3}$$

H is the new height parameter.

The procedure outlined above is equivalent to postulating that the simplified waveguide theory does not apply, but by introducing the slight modification that







Fig. 3—Variation of $h\sqrt{\omega}_r$ with height.

h, a constant in (1), should be replaced by H which varies with frequency, a considerable measurement of agreement is obtained between theory, the experimental results, and what is known of the constants of the ionosphere. Physically the modification seems plausible. As indicated by Budden⁶ one would anticipate the penetration of the fields of the waveguide modes into the ionosphere to be greater the larger the wavelength, and thus (2) and (3) for the variation of H with f are not unreasonable. Again, the two layer model of Wait⁷ shows that frequencies from 8 to 18 kc are effectively reflected from the lower layer, while frequencies less than 3 kc penetrate to the higher level. Indeed, (2) and (3) may be regarded as an application to frequencies below 1 kc of an extension of the Wait two layer model; the model is now infinitely layered; that is, continuous. Eq. (3) is reminiscent of formulas associated with the skin effect, but too much significance should not be ascribed to this similarity.

⁶ A. H. Waynick, "The present state of knowledge concerning the lower ionosphere," Proc. IRE, vol. 45, pp. 741-749; June, 1957.

⁶ K. G. Budden, "The propagation of very low frequency radio waves to great distances," *Phil. Mag.*, vol. 44, pp. 504-513; May, 1953.

⁷ J. R. Wait, "An extension to the mode theory of VLF ionospheric propagation," *J. Geophys. Res.*, vol. 63, pp. 125–135; March, 1958.

The Propagation of Radio Waves of Frequency Less Than 1 kc*

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Summary-The simplified mode theory of propagation in a waveguide formed by the earth and a concentric ionosphere of constant height is applied to the experimental observations of Chapman and Macario for the frequency range between 100 cps and 1000 cps. It is demonstrated that the discrepancies between the theory and the nighttime experimental results may be explained by modifying the theory and postulating an effective increase in the ionospheric height as the frequency decreases. This concept is also shown to be not necessarily incompatible with the results for day.

INTRODUCTION

NCREASING attention has been paid of late to propagation at extra low frequencies (ELF). At these frequencies it appears that the propagation can be very simply explained in terms of a single mode travelling within the waveguide formed by the earth and a concentric homogeneous ionosphere at height *h*.

According to the mode theory of propagation in the form developed by Wait,^{1,2} only the zero mode is of importance at ELF. After certain approximations, the attenuation coefficient α in this mode can be expressed as:

$$\alpha \approx \frac{7850}{h} \sqrt{\frac{f}{\omega_r}} \quad \text{where} \quad \omega_r = \frac{Ne^2}{m\epsilon_0 \gamma} \cdot \qquad (1)$$

In (1), α is measured in decibels per 1000 km, h is in kilometers, f is the frequency, ω_r is in mks units with N and γ the electron density per m^3 and the electron collisional frequency, respectively, for the homogeneous ionosphere; the other symbols have the conventional significance. It is evident from (1) that if the simple theory applies, and h and ω_r are therefore not to be considered as dependent upon the frequency of the wave being propagated, then $\sqrt{f/\alpha}$ should be constant.

THE EXPERIMENTAL INFORMATION

Experimental values for the attenuation coefficients at ELF are regrettably scarce. The only strong natural source of signals at these frequencies is the lightning discharge, and it is from studies of the propagation of atmospherics that Chapman and Macario³ have derived

figures for the attenuation coefficients. Some values have also been obtained by Holzer, Deal, and Ruttenberg,⁴ but these have not vet been published in the general literature; even the valuable information of Chapman and Macario is only available in summarized form and not in the detail that could be desired. It should perhaps be pointed out that the single station work of Chapman and Macario is not an entirely satisfactory source of data. For instance, the observations are limited to distances comparable with the wavelengths at the lower range of ELF; this implies that near field analysis should be applied. Again in single station work, as compared with investigations using several stations, the variable spectral content of the source cannot be eliminated, and it is therefore necessary to average many individual results. This procedure has both advantages and disadvantages. The conclusions drawn can never be as precise as those obtained from multi-station recording on the same atmospheric. On the other hand it is notoriously misleading in geophysics, where many quantities show considerable variations, to concentrate upon an individual example and ignore the over-all behavior.

Chapman and Macario's experimental results do not appear to have been examined on the basis of (1). Accordingly, this has been done in Fig. 1 in which two curves are plotted of \sqrt{f}/α against f for day and night conditions respectively, and in the frequency range of 100 cps to 1000 cps. It is apparent that during day for frequencies exceeding about 300 cps the constancy of \sqrt{f}/α to be expected on the simple theory is realized; at night, there is no indication of such a constancy; and both by day and by night there are divergencies for the lower frequencies. The last effect is perhaps not entirely unexpected, since (1) is really only valid for distances exceeding about one or two wavelengths, and, as already indicated, the experimental observations are limited to within 3000 km (wavelength for 100 cps). Again at distances "close" in terms of a wavelength, the electrostatic component of the field due to a lightning discharge is significant. Indeed, a combination of these influences would imply that the short range attenuation law should differ from that for greater distances. There is some suggestion of this in the results of Chapman and Macario, but the spread of the observations is too great for any precise conclusions to be drawn.

^{*} Original manuscript received by the IRE, May 4, 1959; revised manuscript received. December 1, 1959 † AVCO Research and Advanced Development Division, Wil-

mington, Mass. J. R. Wait, "The mode theory of VLF ionospheric propagation

for finite ground conductivity," PROC. IRE, vol. 45, pp. 760-767;

<sup>June, 1957.
² J. R. Wait, "The attenuation vs frequency characteristics of VLF radio waves," PROC. IRE, vol. 45, pp. 768–771; June, 1957.
³ F. W. Chapman and R. C. V. Macario, "Propagation of audio-frequency radio waves to great distances,"</sup> *Nature*, vol. 177, pp. 930–932. Nucl. 1056. 933; May, 1956.

⁴ R. E. Holzer, O. E. Deal, and S. S. Ruttenberg, "ELF propa-gation," *Proc. VLF Symp.*, Boulder, Colo., Paper No. 45; January, 1957.

a minimum return loss of 15 db from 50 to 60 kmc has been developed but has not yet been incorporated into a tube. With such transducers flatter characteristics will be obtained.

SCALING TO LOWER AND HIGHER FREQUENCIES

It may be interesting to speculate on the uses of the techniques developed for this tube and extensions of these techniques in high-power tubes of lower frequency. Thus, suppose we scale the present tube down in frequency to X-band. Maintaining γa , ka, and the current density constant, and allowing for the decreased attenuation at the lower frequency, one watt at 55 kmc would correspond to 72 watts at 10 kmc. To cool the helix of such a tube successfully, techniques of improving the heat transfer across the boundary between the dielectric rod and the heat sink would have to be developed beyond those used in the mm-wave tube. The CW output at X-band might be further increased to over 200 watts, at the expense of a higher voltage, by increasing ka to 0.4. As discussed before, we would be depending on the stop band at a ka of 0.5 to prevent backward-wave oscillations.

As for going to higher frequencies, it appears that an increase in ka to 0.4 (while maintaining the helix diameter the same) might permit operation at 80 kmc. Operation at a ka of 0.65—*i.e.*, above the stop band but below the first forbidden region of the helix—might permit us to go to 150 kmc. Both of these tubes would have CW output power in the range of about 100 milliwatts.

Conclusion

The results presented in this paper are of a preliminary nature. However, we have accumulated sufficient experience with this tube to convince us that there will be no major stumbling blocks between the point we have now reached and our ultimate objective of a longlife tube with refined and reproducible performance. By the use of the mechanical techniques described here it is possible to keep tolerances reasonably liberal even though we obtain final alignments accurate to within tenths of a mil. It should therefore be possible to construct this tube in quantity and at reasonable cost. Thus we feel that we have demonstrated the practicality of a broad-band amplifier capable of delivering power outputs of up to one watt in the 5- to 6-mm band.

Acknowledgment

The millimeter wave amplifier described here has benefited importantly from contributions of a number of colleagues. L. J. Speck developed the means of constructing the tiny helix which is the heart of the tube. He also developed the mica vacuum windows and was responsible for the execution of the over-all mechanical design. K. E. Schukraft of the Murray Hill Precision Room developed the technique of grinding copper which we have used extensively in the construction of the tube. G. F. Herrmann designed the electron gun and P. P. Cioffi developed the magnetic circuits. M. E. Hines made valuable suggestions with regard to the support rod and to electrical contact to the helix.

CORRECTION

H. A. Wheeler, author of "The Spherical Coil as an Inductor, Shield, or Antenna," which appeared on pages 1595–1602 of the September, 1958 issue of PRO-CEEDINGS, has requested that the following corrections be made to his paper.

Formula (29) on page 1600 should read:

$$p = \frac{R}{\omega L} = \left(\frac{2\pi a}{\lambda}\right)^3 \frac{1}{1+2/k} = \frac{V}{V_r} \frac{1}{1+2/k} \cdot \frac{1}{V_r}$$

Formula (31) on page 1600 should read:

$$\frac{3\pi^2 a' \delta^2}{2\lambda}$$

to agree with formula (3) of the author's earlier paper, reference [18].

Formula (46) on page 1602 should read:

$$n = \left(\frac{27L^2}{2\pi\mu_0^2.4}\right)^{1/3}$$





Fig. 11-Power output as a function of beam current.

periments the beam current was changed by varying both the anode and beam forming electrode voltages. The saturation power output, both before and after fading, was measured; Fig. 11 shows the results. It is seen that a power output of one watt at 5-ma beam current was obtained, but that the tube could not hold this level and faded back to about one-half watt. As discussed earlier, this fading was caused by heating of the output end of the helix by RF dissipation and the con-





Fig. 13—Power output as a function of frequency.

sequent increase in RF dissipation. Even at the onewatt level there was little increase in helix interception. The power output before fading is approximately proportional to the four-thirds power of beam current as expected. It should be noted that this tube had F-66 ceramic support rods, and that with the better thermal conductivity of sapphire it may be possible to maintain a considerably higher output level.

The low-level gain and the saturation power output over the band are shown in Figs. 12 and 13 for a beam current of 3 ma. The detailed shapes of these curves are largely determined by the helix-to-waveguide transducers initially used. A much improved transducer with



Fig. 7—Structural features of the millimeter-wave TWT (artist's rendering).

waveguide windows are made as separate subassemblies which can be individually leak checked and tested for their microwave properties. They consist of a one-mil thick synthetic mica flake glazed to a kovar cup which, in turn, is brazed across the copper waveguide. The windows have a transmission loss of about 1 db and VSWR of about 2 db. Once a window has been found satisfactory, it is fastened against its seat in the matching block by a heliarc weld. The matching sections are located concentric with the helix by optical alignment and are then screwed to the helix block. (For simplicity these screws are not shown in Fig. 7.) The outside diameters of the pole pieces are thus concentric with the helix and can be used for aligning the tube in the magnetic field. Finally, the four-gun positioning studs are located on a circle concentric with the helix-again by optical means.

The gun assembly is shown schematically (and not to scale) on the left in Fig. 7. A subassembly of three parallel platforms glazed to four ceramic rods is first prepared.' Electrodes are then placed on these platforms one at a time, aligned optically, and screwed into place. The complete gun is finally mounted on the input matching section, and positioned concentric with the helix by the studs.

The tube is completed by adding collector and stem, and by closing the vacuum envelope by means of heliarc welds. Throughout the design, care was taken in the selection of materials to insure that during thermal cycling parts did not shift as a result of differential expansion pressures. Fig. 8 shows a photograph of the completed tube and Fig. 9 shows the tube mounted in a permanent magnet. This magnet provides 1500-gauss axial field with the transverse component held to less than 1/10 per cent. The magnetic pole pieces inside the tube serve as part of the magnetic circuit and thus shield gun and collector from magnetic fields. The holes in the magnetic pole pieces determine the magnetic axis of the circuit; because they are concentric with the helix they make the magnetic axis coincident with the helix axis.



Fig. 8---Photograph of the millimeter-wave TWT.



Fig. 9—Photograph of the tube mounted in a permanent magnetfocusing circuit.

Since the gun has been carefully constructed to produce a beam which is accurately coaxial with the helix, all three critical axes—beam, helix and magnetic field have thus been made coincident, and very good focusing has been consistently obtained without the need for tedious final alignment procedures.

EXPERIMENTAL RESULTS

A total of seven tubes have given RF output powers of 100 milliwatts or more in the 5–6-mm band. Results will be presented for one of the later tubes in which several of the early problems have been eliminated. Fig. 10 shows typical curves of RF power output as a function of input power at midband for a beam current of 3 ma. These curves were taken with the helix voltage adjusted for maximum gain at low signal levels. Maximum power output at saturation is obtained at only very slightly higher helix voltages, and the resultant characteristics appear almost identical to the ones shown here.

Experiments were performed to determine the maximum output power capability of the tube. In these ex1960

value results first from the high-voltage design which results in low C, and second, from the effect of attenuation which reduces the efficiency by a factor of about 3 over that which could be obtained at low frequencies. Presumably the over-all efficiency of the tube could be raised somewhat by low-voltage collection. For the low QC value (0.07) of this tube, Cutler's experiments predict a small enough velocity spread so that collection at as low as 0.3 of the helix voltage should be feasible.

Mechanical Design

The very small internal diameter of this helix (15 mils) and its very large length-to-diameter ratio give rise to a tube which requires a high degree of precision in its construction. Helix straightness and the alignment of gun, helix, and magnetic field axes must be maintained within a tolerance of about one-half mil over the four-inch length (i.e., within an accuracy of 1 in 8000). To obtain such precision through the use of self-aligning piece parts would have required so many tight tolerances as to render the tube impractical. The problem of alignment may be viewed as two-fold: first, the concentric alignment of parts, and second, the angular alignment of their axes. We chose to obtain concentric alignment by use of optical techniques, and angular alignment by the use of a number of parallel reference surfaces. This simplified the design to the point where we could take full advantage of machining operations in which high precision is obtained relatively easily. Examples of such operations are surface grinding to produce flatness and parallelism, centerless grinding to produce constancy of diameter (although not necessarily straightness), and machining with a single lathe setting to produce concentricity of cylindrical surfaces. By extensive use of these techniques we have achieved alignment of gun, helix and magnetic field axes to within the required one-half mil without imposing tolerances tighter than one mil on absolute dimensions. By contrast, the use of precise interlocking piece parts would have required tolerances tighter than one-tenth mil to give this precision of alignment.

The helix assembly is shown in Fig. 5. The helix is wound from 2×4 mil molybdenum tape, glazed to a single wedge of dielectric, and then copper plated. This single-rod support has several advantages. First, it gives rise to structural simplicity and avoids the problem of stress due to mechanical overconstraint. Second, it minimizes the amount of dielectric in the RF fields and thus minimizes both dielectric loading and dielectric loss. Finally, the perturbation (once each turn) due to the dielectric rod introduces a stop band into the helix transmission around the frequency for which ka = 0.5. This helps to suppress any tendency toward backwardwave oscillation.

The method of supporting the helix is shown in Fig. 6. The ceramic support rod is forced into the corner of a copper block by a multiple-finger spring which contacts



Fig. 5 – Photograph of the belix glazed to dielectric rod (helix pitch is 110 TP1).



Fig. 6-Helix mounted on a heat sink (artist's rendering).

the rod along the entire length of the helix. This block serves as a heat sink and has a direct heat conduction path of low impedance to the outside of the vacuum envelope. The surfaces of the block against which the helix is mounted are ground accurately straight, but need not be accurately located. The spring and the ceramic rod are so designed that there is a net component of force pushing the rod flat against the bottom surface and back into the corner. The dielectric rod is anchored to the helix block at its midpoint only and is otherwise free to slide axially with respect to the copper block as required by differential expansion during bake-out. The end surfaces of the block are ground perpendicular to the two planes which form the corner against which the helix is mounted, and thus serve as precise reference surfaces for aligning the gun and magnetic field axes with respect to the helix.

The input and output matching sections which contain the waveguide-to-helix transitions, vacuum windows, and means for supporting gun or collector are mounted as shown in Fig. 7. These matching sections are brazed stack-ups of two copper disks and a steel disk. The copper disks contain the waveguide, and the steel disk serves as a magnetic pole piece. After brazing, the flat surfaces are ground accurately parallel, as suggested in Fig. 7, and the inside hole is machined accurately concentric with the outside diameter. The It is the space charge term which plays a dominant role in the much discussed case of Brillouin focusing. However, as the diameter of a beam is made smaller and smaller, the motion associated with transverse emission velocities remains essentially constant while the transverse motion due to space charge decreases with decreasing beam radius. In our tiny beams it is therefore not surprising that the magnetic field term due to thermal velocities is about $2\frac{1}{2}$ times as large as that due to space charge. The actual dimensions of the millimeter tube gun are shown in Fig. 3. The gun is completely shielded from the magnetic field, making the third term in the magnetic field expression zero.

For a given beam diameter and current, the thermal velocity term B_t^2 is proportional to cathode area. Consequently, to minimize the required field, the cathode current density was made as high as possible, consistent with the probability of long cathode life.

A relatively long gun with a small angle of convergence was used to keep lens effects small. The measured beam convergence on scaled-up models of this gun has been about 20 per cent greater than calculated. To reduce the convergence and optimize focusing in the actual gun, a positive bias of 5 volts is applied to the beam-forming electrode. The accelerating anode voltage must therefore be reduced from its design value in order to maintain the desired beam current. This necessitates post acceleration between the anode and the helix, thus giving rise to an additional lens effect. The magnetic field is introduced rather suddenly by an aperture in a magnetic pole piece brought inside the vacuum envelope. No attempt has been made as yet to optimize the geometry of this region. Experimentally, about 95 per cent beam transmission from cathode to collector has been obtained with 1500 gauss magnetic field. This compares favorably with a calculated minimum field of 1200 gauss. With some effort at eliminating the postacceleration and optimizing the entrance conditions of beam into the magnetic field, it should be possible to increase the transmission.

ATTENUATION AND EFFICIENCY

At millimeter wavelengths, helix attenuation becomes sufficiently high to have a major effect on power output. Fig. 4 shows the expected relationship between these quantities. We have plotted the ratio of efficiency to the gain parameter C as a function of the ratio of L/Cwhere L is the loss per wavelength. The intercept of the curve for zero attenuation is obtained using results of an experimental study of TWT efficiency by Cutler,¹² and the slope is determined from another experimental study of the effect of loss on efficiency by Cutler and Brangaccio.13

 ¹² C. C. Cutler, "Nature of power saturation in traveling-wave bes," *Bell Sys. Tech. J.*, vol. 35, pp. 841–876; July, 1956.
 ¹³ C. C. Cutler and D. J. Brangaccio, "Factors affecting traveling ve tube power capacity," IRE TRANS. ON ELECTRON DEVICES, NUMBER OF COMPARISON OF COMPARI tubes.

wave tube power capacity," II vol. ED-3, pp. 9-23; June, 1953.



Fig. 3-Electron gun. (All dimensions in inches.)



Fig. 4---Approximate relationship between traveling-wave tube efficiency and L/C

For a molybdenum helix of the type used in the millimeter tube the L/C ratio as based on actual measured helices would be about 25. According to Fig. 4 this would result in an efficiency of about C/3. To reduce the L/C ratio we copper-plate the helix. With reasonable care in the plating process, we can obtain a plated surface with 0.8 the conductivity of solid copper. This would reduce the L/C ratio to about 12 and the resulting efficiency would be about 0.9 C. By plating the helix we will gain somewhat more than the factor of about 3 indicated here, since the less lossy helix will in turn dissipate less RF power. This means that the power fade caused by RF heating will be somewhat less. With a C value of 0.015 this gives us an expected electronic efficiency of about 1.4 per cent. This comparatively low

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Fig. 2—Helix temperature vs power dissipated (per inch) in helix structure.

At the very output end, the helix will also be heated by RF power dissipation. This heating causes an increase in helix attenuation as a result both of increased wire resistivity and increased dielectric loss. The increased attenuation, in turn, lowers the power output thereby giving rise to fading; *i.e.*, if the beam is turned on suddenly, the power output will go to a maximum value and then decrease gradually over a period of about one minute while the helix heats up. The fading problem is aggravated by the necessity of stopping the heat sink short of the last few turns of the helix to make room for the helix-to-waveguide transducer. This makes the heat flow path from the end turns to the heat sink longer than from the remainder of the turns and thus increases the fraction of the total thermal impedance which is due to the dielectric rod. Thus, to minimize fading, it is especially important to obtain a high thermal conductivity dielectric rod. Fading has limited the power output of tubes with steatite rods to about one-half watt CW. The use of sapphire with about ten times the thermal conductivity will reduce fading considerably. Beryllium oxide which offers an additional factor of ten in thermal conductivity would probably produce little further improvement in fading in the present tube. However, if means were found to reduce the thermal impedance between dielectric rod and heat sink, this material should make possible a considerable increase in beam power. Coupled with an increase in ka to 0.4, this could conceivably result in a CW output of several watts.

BEAM FOCUSING

The problem of focusing the tiny electron beam differs from that commonly encountered in that the transverse thermal velocities with which electrons are emitted from the cathode become the predominant factor in determining the magnetic field required for adequate focusing.

To achieve a high-voltage convergent beam of much smaller diameter than had been produced in any of our previous work on convergent TWT beams, the gun design effort was accompanied by theoretical beam studies of quite general application. Herrmann,¹⁰ who carried out both the analytical and the experimental phases of the initial gun work, extended calculations of thermal velocity effects in electron beams beyond the magnetically shielded gun region¹¹ to include the full region of magnetic focusing between accelerating anode and collector. Two particularly significant conclusions drawn from Herrmann's study are as follows:

- Where thermal velocities play a dominant role, minimum magnetic field is obtained by shielding the cathode from the field.
- 2) Given the desired beam parameters (voltage, current and radius) and the cathode current density, the minimum magnetic focusing field can be predicted without any knowledge of the specific gun geometry used. Comparison of the actual focusing field required with this theoretical minimum thus affords a measure of excellence of the entire focusing system.

The expression for the minimum field is made up of three terms as follows:

$$B^{2} = B_{b}^{2} + B_{t}^{2} + B_{k}^{2} \left(\frac{A_{k}}{A_{b}}\right)^{2}$$

Space Thermal Cathode
Charge Velocities Flux

where

- B = minimum total magnetic flux density required,
- B_b = Brillouin field to counteract space charge forces,
- B_t = field required to counteract spreading due to thermal velocities,

 B_k = magnetic flux density at the cathode,

- $A_k = \text{cathode area, and}$
- $A_b =$ area of the beam.

¹⁰ G. F. Herrmann, "Optical theory of thermal velocity effects in cylindrical electron beams," *J. Appl. Phys.*, vol. 29, pp. 127–136; February, 1958.

¹¹ The basic approach to calculations in the accelerating region of a shielded Pierce gun had been developed by C. C. Cutler and M. E. Hines, "Thermal velocity effects in electron guns," Proc. IRE, vol. 43, pp. 307–315; March, 1955. More detailed calculations applicable to shielded guns having a larger range of thermal velocities had been given by W. E. Danielson, J. L. Rosenfeld, and J. A. Saloom, "Detailed analysis of beam formation with electron guns of the Pierce type," *Bell Sys. Tech. J.*, vol. 35, pp. 375–420, March, 1956; also by G. F. Herrmann, "Transverse scaling of electron beams," *J. Appl. Phys.*, vol. 28, pp. 474–478, April, 1957. to be described. At the time, a ka of 0.25 seemed about as high as we could safely go without incurring backward wave oscillations. Subsequently we found that our method of supporting the helix introduced a strong stop-band around ka = 0.5.⁹ This means that we could probably increase ka to 0.4 in a future redesign without incurring backward wave oscillations. To make the frequency response of this tube as flat as possible, we chose γa equal to 1.5. This, together with the choice of ka, led to a comparatively high beam voltage and a large helix-length-to-diameter ratio. The relevant tube parameters are as follows.

 $\begin{array}{c} ka = .25\\ \gamma a = 1.5 \end{array}$ at 55 kmc Helix inside diameter = 15 mils Helix pitch = 110 turns per inch Helix length = 4 inches Synchronous Voltage = 7000 volts Beam current = 3 ma Gain parameter C = 0.015

The values of helix length and beam current were calculated assuming an output of 100 mw, an efficiency of one times C and a low-level gain of 25 db. Some of these parameters are discussed in more detail later.

HEAT TRANSFER FROM THE HELIX

The helix is heated both by RF dissipation and intercepted beam current. This heat must be removed while maintaining the helix temperature low enough to prevent an undue increase in RF attenuation and a consequent loss of efficiency. This we have done by providing a direct thermal conduction path from the helix wire to the outside of the vacuum envelope. As shown in Fig. 1, the helix is glazed to a single wedge of dielectric which is forced against a massive copper heat sink by strong spring pressure. Dielectric rods of either Bell Laboratories F-66 steatite or of synthetic sapphire have been used. Beryllium oxide appears to be a still better material because of its very high thermal conductivity and it will be tried in the future. In finely divided form, however, it is highly toxic, and, its experimental evaluation will have to await the development of special grinding and processing techniques. These three dielectric materials-steatite, sapphire, and beryllium oxidehave thermal conductivities in the approximate ratio 1:10:100.

There are three sources of thermal impedance in the structure of Fig. 1. These are the glazed joint between helix and dielectric rod, the rod itself, and the interface between rod and heat sink. The first of these is rather

ing that conduction is the main mechanism of heat transfer and that the thermal impedances are roughly constant. Whereas the intrinsic thermal conductivities of steatite and sapphire differ by a ratio of ten, the

curves show that the apparent conductivities for the two materials differ by a factor of only 3: 200°C/watt/ inch-of-helix-length for steatite vs 65°C/watt/inch for sapphire. This means that the thermal impedance of the interface between dielectric rod and heat sink is not negligible. Assuming it to be the same for the two materials, as a first-order approximation, we calculate its value to be 50°C/watt/inch. The impedance of the dielectric rod is then 150°C/watt/inch for steatite and is 15°C/watt/inch for sapphire. This shows that for the case of sapphire the interface impedance will largely control the temperature rise of the helix. A rough calculation of the termal impedance of the steatite rod from its geometry shows good agreement with the above results.

For comparison, Fig. 2 also shows the temperature rise for a helix glazed to a ceramic rod but suspended without contact to the heat sink so that it can only be cooled by radiation. As may be seen, the power that can be radiated from the helix for a given temperature rise is very much smaller than that which can be carried away by conduction.

Most of the helix heating is caused by beam bombardment-RF heating occurring in any appreciable magnitude only along the last few turns near the output. The beam interception in experimental tubes has ranged from 5 to 7 per cent. Assuming that this interception is distributed over one quarter of the helix length, we find that the temperature rise is 300°C for a steatite rod helix and 100°C for sapphire. Both of these values are within nermissible limits.

Fig. 1—Cross section of helix mounted on heat sink.

small because the glaze, in wetting both the helix and

the rod, provides an intimate thermal contact between

them. The other two impedances may, however, be appreciable. We obtain a measure of their relative magni-

tudes by comparing the power dissipation properties of

helices with steatite and sapphire rods. This was done

by passing dc current through some sample helices

mounted on a heat sink in a vacuum bell jar. The helix

temperature was calculated from the rise in wire re-

sistance and plotted against the dc power dissipated. Fig. 2 shows the results. The curves for helices mounted

on a heat sink are approximately straight lines indicat-

⁹ This means of introducing a stop band is similar in principle to an earlier arrangement used by Poulter in which a helix was mounted in a quartz tube having a groove which intercepted the helix once each turn. See W. L. Rorden, "A 100 Watt CW TWT at S Band," Electronics Research Lab., Stanford Un versity, Tech. Rept. 351-1; March 9, 1956.

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A Half-Watt CW Traveling-Wave Amplifier for the 5-6 Millimeter Band*

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Summary-The traveling-wave tube here described represents the first practical CW power amplifier with broadband performance in the millimeter wave band. More than 30 db of gain has been achieved over a bandwidth of 10,000 mc centered at 55,000 mc. A maximum CW output power of 1 watt has been obtained in this band. This combination of high power output and broadband performance represents a significant advance in the millimeter art.

The electrical and mechanical techniques are described which were found successful in solving the problems peculiar to the high operating frequency. These problems are focusing, heat dissipation, intrinsic RF loss and structural precision. Experimental data on an operating tube are presented which show good agreement with predicted performance. These data suggest that the techniques described may be extended to allow either an increase in power to several watts at the present operating frequency or an increase in operating frequency to 150 kmc for output powers of a hundred milliwatts.

INTRODUCTION

OR SOME years now the possibility of broadband communications at millimeter waves has been studied at Bell Telephone Laboratories.¹ This study has centered around the use of circular waveguide as a low-loss transmission medium. Transmission over a 40-kmc wide band, from 35-75 kmc, appears feasible in a single waveguide, but initial interest has been focused on the 50 to 60 kmc region. This interest in millimeter wave communications has stimulated vacuum-tube work both on primary signal sources²⁻⁵ and on amplifiers. In this paper we shall describe an experimental helix-type traveling-wave amplifier with a CW power output of $\frac{1}{2}$ watt in the 50 to 60 kmc band. This exceeds output powers previously obtained from CW amplifiers in this frequency range by at least an order of magnitude.

Results on early work on millimeter wave amplifiers were reported by Little⁶ in 1951. In his tube a tiny unsupported helix was stretched between two posts and flooded by an electron beam. Cooling was entirely by radiation and output power thereby limited to a few

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 SFD Labs., Inc., Union, N. J. Formerly at Bell Telephone Labs., Inc., Murray Hill, N. J. ‡ Bell Telephone Labs., Inc., Murray Hill, N. J.

¹ S. E. Miller, "Waveguide as a communication medium," *Bell Sys. Tech. J.*, vol. 33, pp. 1209–1265; November, 1954.
 ² E. D. Reed, "Tunable low voltage reflex klystron for operation in the 50–60 KMC band," *Bell Sys. Tech. J.*, vol. 34, pp. 563–599;

May, 1955. ³ A. Karp, "Traveling-wave tube experiments at millimeter wavelengths with a new easily built space harmonic circuit," PROC. IRE, vol. 43, pp. 41–46; January, 1955.

vol. 43, pp. 41-40; January, 1955.
⁴ A. Karp, "Backward wave oscillator experiments at 100-200 KMC," Proc. IRE, vol. 45, pp. 496-503; April, 1957.
⁶ C. F. Hempstead and A. R. Strand, "Versatile source of millimeter waves," *Bell Labs. Rec.*, vol. 35, pp. 241-245; July, 1957.
⁶ J. B. Little, "Amplification at 6-millimeter wavelength," *Bell Labs. Rec.*, vol. 29, pp. 14-17; January, 1951.

microwatts. The net gain of this tube was about 3 db at 6 mm. Also in 1951, Millman⁷ reported 20 db of low-level gain and an output power of about 20 mw at 7 mm from a tube having an all-metallic filter-type circuit. This tube had a bandwidth of about 7 per cent. In later millimeter wave work, the helix-type tube was revived by Robertson⁸ in an attempt to obtain wider bandwidth. He supported the helix by four knife edges of quartz, but again cooling was mainly by radiation. He obtained a pulse power output of 5-10 mw at a 5 per cent duty cycle. The net gain was 10-15 db at 6 mm.

At the start of this program we thought that heat dissipation associated with the relatively high power output would force us to use an all-metallic filter-type structure such as Millman's, with a consequent loss in bandwidth. However, further studies showed that adequate power output could be obtained with a helix, if it was cooled by conduction and interception was kept very low. These studies led to the helix-type tube described below.

DESIGN CONSIDERATIONS

The major problems encountered in the development of this amplifier were direct consequences of the very small helix diameter necessitated by the high operating frequency. These problems were:

- 1) the need for adequate cooling of the helix;
- 2) the production of an electron beam of very small diameter and the focusing of this beam through the helix without requiring excessive cathode current density or excessive magnetic field;
- 3) the minimization of intrinsic helix attenuation; and
- 4) the precise alignment of gun, helix and magnetic field.

Our solutions to these problems will be treated separately in the following sections.

Direct scaling of typical lower-frequency TWT's would have resulted in a prohibitively small helix diameter of about 5 mils. To enable us to use a larger diameter, we chose to operate at the high ka value of 0.25. As a result we were able to use a 15-mill diametera value which seemed attainable through the techniques

⁷ S. Millman, "Spacial harmonic traveling-wave amplifier for six millimeters_wavelength," PROC. TRE, vol. 39, pp. 1035–1043; Sep-

^{*}S. D. Robertson, "Broadband helix traveling-wave tube for millimeter wavelengths," IRE TRANS. ON MICROWAVE THEORY AND TECHNIQUES, vol. MTT-2, pp. 48-54; September, 1954.

could arrive at the conclusion that the surface potentials of these three units should be in the order $\phi_s(\text{E-3}) \ge \phi_s(\text{E-1}) \ge \phi_s(\text{E-2})$.

The authors believe that in certain cases the information provided by the data in Fig. 3 may permit the computation of surface recombination velocity. This was done for transistor E-2 as follows.

If one assumes that *s* is made negligibly small when a negative voltage is applied to the field plate facing the emitter of transistor E-2, the surface recombination velocity term in the Webster equation may be neglected for that transistor, and an estimate may be made of the value of bulk lifetime, since the constants left in the equation may be obtained from independent measurements. Using the computed value of τ_b , s may then be estimated for transistor E-2 when no voltage is applied to the field plate (see Appendix). The surface recombination velocity was found to be 500 cm/sec and a displacement of 1.7×10^{12} charges/cm² increased s by 2000 cm/sec for this particular transistor. The computed value of s for transistor E-2 was rather small for silicon, but E-2 did have a relatively high α , which would imply a low surface recombination velocity.

CONCLUSION

The surface potential of an operating silicon transistor may be estimated by means of the field effect measurement if a strong enough field can be achieved. If this field strength can be reached, the surface recombination velocity may also be computed.

The time constants obtained for the $\Delta \alpha$ decay are dependent upon the age of the surface of the device and may give valuable information about the oxide layer covering the silicon surface, as well as about the mechanism of loss of carriers at the surface.

The field effect may also be used for determining optimum geometry configurations in an operating transistor.

Appendix

The following is Webster's equation which relates the power gain of a transistor to surface recombination velocity s, bulk lifetime τ_b , and emitter current I_e .

$$\frac{1}{\alpha} - 1 = \frac{sA_sW_b}{2AD_p} + \frac{W_b^2}{2D_p\tau_b} + \frac{W_b^2\mu_e}{2A\sigma_e L_e D_p} I_e, \quad (3)$$

where α is the small signal ac emitter-to-collector current amplification factor, W_b the effective junction spacing, A_* the area over which surface recombination takes place, .1 the entire area of the emitter, D_p the diffusion constant for holes in the base region, L_ϵ the diffusion length for electrons in the emitter region, μ_ϵ the electron mobility, and σ_ϵ the conductivity of the emitter region adjacent to the junction.

For the transistors used in these experiments

$$W_b \cong 2.54 \times 10^{-3}$$
 cm,

$$\mu_e = 48 \frac{\mathrm{cm}^2}{\mathrm{sec \ volt}},$$

$$I_e = 1 \times 10^{-3} \text{ amperes},$$

$$A = 1.26 \times 10^{-2} \mathrm{cm}^2,$$

$$\sigma_e = 56 \mathrm{ohm}^{-1} \mathrm{cm}^{-1},$$

$$L_e = 2 \times 10^{-3} \mathrm{cm},$$

$$D_p = 12 \frac{\mathrm{cm}^2}{\mathrm{seconds}}.$$

Putting in these values for the last term of (3), one gets

$$\frac{1}{2} \frac{W_b^2 \mu_e I_e}{A \sigma_e L_e D_n} = 8.3 \times 10^{-6}.$$
 (4)

This term is negligible in comparison to the sum of the two other terms in (3), since even for a unit with an α of 0.95, $1/\alpha - 1 = 5.3 \times 10^{-2}$.

After the I_e term is neglected, the Webster equation simplifies to

$$\frac{1}{\alpha} - 1 = \frac{W_b^2}{2D_p} \left(\frac{2s}{r} + \frac{1}{\tau_b}\right) \tag{5}$$

if the assumption is made that $A_s = 2 \pi r W_b$, where r is the radius of the emitter junction.

For transistor E-2

 $\alpha_0 = 0.946 = \text{initial } \alpha,$ $\alpha_1 = 0.949 = \text{maximum } \alpha \text{ obtained while applying a}$ negative V to the field plate,

 $\alpha_2 = 0.932 = \alpha$ when +2500 volts are applied to the field plate,

$$W_b = 2.42 \times 10^{-3}$$
 cm,
 $r = 63.5 \times 10^{-3}$ cm,
 $D_p = 12$ cm²/sec.

Assuming s is negligible when α is equal to α_1 , then

$$\tau_b = \frac{W_b^2}{2D_p \left(\frac{1}{\alpha_1} - 1\right)} = 4.5 \times 10^{-6} \text{ sec.}$$
(6)

When no voltage is applied to the field plate, $\alpha = \alpha_0$ and

$$s_{0} = \left[\frac{\frac{1}{\alpha_{0}} - 1}{\frac{W_{b^{2}}}{2D_{p}}} - \frac{1}{\tau_{b}} \right] \frac{r}{2} = 460 \simeq 500 \text{ cm/sec.} \quad (7)$$

When +2500 volts are applied to the field plate, $\alpha = \alpha_2$ and $s_2 = 2500$ cm/sec. Therefore a displacement of 1.7×10^{12} electrons/cm² increases s by 2000 cm/sec for this particular transistor.

Acknowledgment

The authors wish to thank M. Cutler, H. Bath, and R. Solomon for their many helpful comments and suggestions.

and negative fields. These results led to the conclusion that if the transistor is operated in the normal direction, the size of the collector is large enough so that essentially no minority carriers are lost at the surface surrounding the collector.

INTERPRETATION OF RESULTS

The authors believe that the data of Fig. 3 may be used to determine qualitatively the surface potential of the transistors used. Webster's equation¹⁷ indicates that α is essentially inversely proportional to surface recombination velocity, *s*. When a field is applied normal to the surface of a transistor the surface potential is changed, and in turn surface recombination velocity is affected. Many, *et al.*,¹⁸ have shown a relationship between *s* and ϕ_s which is plotted in Fig. 5 for a trapping state energy of 0.48 ev, as found by Statz, *et al.*,¹⁹ and a capture probability ratio of 5×10^8 , which was reported by Buck and McKim for silicon.²⁰

One can qualitatively place the surface potential of the three transistors of Fig. 3 at the positions shown in Fig. 5. This is done by observing in Fig. 3, for example, that for transistor E-2 the $\Delta \alpha_i$ saturates at about -1200volts. When transistor E-2 is placed at the position shown in Fig. 5, it is seen that *s* "saturates" at its lowest value when ϕ_s is made more negative, and does not saturate when ϕ_s is made more positive. These factors are in good agreement, since a negative voltage on the field plate induces the surface potential to become more negative.

The placement of the three transistors in Fig. 5 merely from the data of Fig. 3 results in positive surface potentials for transistors E-1 and E-3. This raises what appears to be a disturbing contradiction, since the results of microlight-probe measurements on devices of the same type as those used here have always shown, under the same conditions of fabrication, treatment, aging, and measurement under standard atmospheric conditions, a conversion of type at the surface. The apparent contradiction can be overcome by consideration of two factors. First, if the capture cross section ratio were reduced by about one order of magnitude, then the placement of transistor E-1 could be such as to result in a conversion of type at the surface. At the same time, this would cause transistor E-2 to shift in surface potential into a region which agrees qualitatively with values obtained in some preliminary measurements of surface

¹⁰D2; 1957.
¹⁹ H. Statz, G. A. de Mars, L. Davis, Jr., and A. Adams, Jr.,
"Surface states on silicon and germanium surfaces," *Phys. Rev.*, vol. 101, pp. 1272–1281; February, 1956.
²⁰ T. M. Buck and F. S. McKim, "Effect of contain channel tract.

²⁰ T. M. Buck and F. S. McKim, "Effect of certain chemical treatments and ambient atmospheres on surface properties of silicon," *J. Electrochem. Soc.*, vol. 105, pp. 709–714; December, 1958.



Fig. 5—Normalized surface recombination velocity vs surface potential (in electron volts).

potential now being performed.21

The second factor is the effect of a vacuum on ϕ_s . Since transistor E-3 was measured not under atmospheric conditions but rather at a vacuum of 10⁻⁴ mm of Hg, special consideration must be given this case. When this unit was measured under atmospheric conditions the properties of this transistor looked very similar to those of transistor E-1; that is, there did not appear to be any saturation of $\Delta \alpha_i$ as the field electrode voltage was made positive or negative. When the unit was placed in a vacuum, the α_0 decreased noticeably and the emitter diode reverse current also decreased. These effects could have been produced by an increase in ϕ_{s} ,²² which would allow for the surface to reach a value of ϕ_s where conversion of type at the surface did not necessarily occur. The lowering of the capture cross section ratio would also be helpful here in that the maximum for *s* could be achieved closer to $\phi_s = 0$.

Another bit of information that agrees qualitatively with the "order" of placement of the three units in Fig. 5 is that the emitter diode reverse currents, as measured at 5 volts bias, were 8, 1300, and 5400 mµa for transistors E-3, E-1, and E-2, respectively. By considering the influence of channels on diode reverse current,²³ one

²³ A. L. McWhorter and R. H. Kingston, "Channels and excess reverse current in grown germanium *p*-*n* junction diodes," PROC. IRE, vol. 42, pp. 1376–1380; September, 1954.

¹⁷ W. M. Webster, "On the variation of junction-transistor current-amplification factor with emitter current," PROC. IRE, vol. 42, pp. 914–920; June, 1954.

¹⁰ ¹⁸ A. Many, E. Harnik, and Y. Margoninski, "Surface recombination processes in germanium and their investigation by means of transverse electric fields" in "Semiconductor Surface Physics," R. H. Kingston, Ed., Univ. of Pennsylvania Press, Philadelphia, pp. 85– 102, 1957.

²¹ Experiments to date indicate that the surface potential of a surface analogous to those used here is about -0.15 volt. This value is based on a very limited amount of information, but does seem to fit the argument qualitatively.

²² Independent experimental measurements by the authors show that a vacuum does produce an increase in ϕ_s .



Fig. 1-Schematic of assembly used in experiments



Fig. $2-\alpha$ vs time while voltage is turned on and off.

Fig. 3 shows what happened to the initial alpha change ($\Delta \alpha_i$) of three transistors when voltages of different magnitude were applied to the field electrode.¹⁶ The initial alpha change of the first transistor was directly proportional to applied positive or negative voltage. This was a low alpha transistor. The initial alpha change of the second transistor was directly proportional to applied positive voltage. However, when negative voltages were applied to the field plate, $\Delta \alpha_i$ appeared to saturate at about -1200 volts. The initial alpha of this transistor was high. $\Delta \alpha_i$ of the third transistor was proportional to applied negative voltage, but when a positive voltage was applied there was no measurable change in alpha.

Fig. 4 is a plot of log $|\Delta\alpha|$ vs time for transistor E-2 operating in the normal direction. A bias of +2600 volts was applied to the field plate which faced the emitter surface. The following equation was obtained from the experimental data:

$$|\Delta \alpha| = .0068e^{-t/60} + .0082e^{-t/340}$$
(1)

and two time constants were required to fit the experimental data.



Fig. 3—Initial $\Delta \alpha$ vs field-plate voltage for three transistors.



Fig. 4—Absolute value of $\Delta \alpha$ vs time for both normal and inverse α of transistor E-2 while a positive voltage was applied to the field plate.

Fig. 4 also shows a plot of log $|\Delta\alpha|$ vs time for the same transistor operating in the inverse direction. Again a positive voltage was applied to the field plate. The field plate faced the same surface as in the preceding experiment; *i.e.*, in both experiments the field plate faced the small area junction surface. In the first case the small area junction was used as the emitter, and in the second case the small area junction was obtained for the latter case from the experimental data:

$$|\Delta \alpha| = 0.063 e^{-t/380}.$$
 (2)

Only one time constant was required to fit the experimental data and, within experimental error, this time constant is approximately the same as the longer time constant of (1). The presence of two time constants in (1) and only one in (2) is a rather confusing point and no completely acceptable argument has yet been developed to explain it.

A similar series of experiments was performed on a transistor where the field plate faced the large area junction. No noticeable effect was observed on normal α . Inverse α was affected only slightly by both positive

¹⁶ Initial alpha change is defined as the maximum observable change in α when a voltage is applied to the field plate. Transistors E-1 and E-2 were measured in room air at 24°C and transistor E-3 was measured in vacuum (10⁻⁴ mm Hg) at 25°C.

Field Effect on Silicon Transistors*

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Summary-A field effect has been observed on an operating silicon transistor. The results have been interpreted as being due to changes in surface recombination velocity produced by changes in surface potential. Using this approach, it was possible to calculate the surface recombination velocity of the base region of an operating device directly, without having to go to a filament measuring technique and then extrapolating back to the device. Relaxation phenomena have also been observed which can be interpreted as being due to a transfer of charge between the fast and the slow states at the surface. Time constants for this transition have been calculated.

INTRODUCTION

THE technique of capacitively applying an electrostatic field to alter the surface properties of semiconductor materials is one that has been used a great deal in the study of surface phenomena. Most of the work reported in the literature concerns itself with measurements made on germanium filaments of a single conductivity type;1-9 some measurements on germanium p-n junctions have also been reported.^{10,11}

With respect to field-effect measurements on silicon, the literature is quite meager. Penman and Brown,¹² Low,13 and Dousmanis,14 have reported on successful

* Original manuscript received by the IRE, June 22, 1959; revised manuscript received, September 21, 1959. † Semiconductor Div., Hughes Products, Newport Beach, Calif.

¹ University of California, Los Angeles. ¹ W. L. Brown, "Surface potential and surface charge distribu-tion from semiconductor field effect measurements," *Phys. Rev.*, vol.

 ¹⁰⁰ pp. 590–591; October, 1955.
 ² H. K. Henisch and W. N. Reynolds, "Surface recombination in the presence of strong electric fields," Proc. Phys. Soc. (London), vol.

B68, pp. 353-356; June, 1955.
 ³ G. G. E. Low, "Modulation of the surface conductance of germanium by pulsed electric fields," *Proc. Phys. Soc. (London)*, vol. B68, pp. 1154–1157; December, 1955.
 ⁴ J. E. Thomas, Jr. and R. H. Rediker, "Effect of electric field on field in the intermediate "*Phys. Rev. Cond.*, 101

surface recombination velocity in germanium," Phys. Rev., vol. 101, pp. 984–987; February, 1956.
 ⁶ A. Many, Y. Margoninski, E. Harnik, and E. Alexander, "Re-

laxation effects in recombination velocity on germanium surfaces under transverse electrostatic fields," Phys. Rev., vol. 101, pp. 1433-1434; February, 1956.

⁶ H. C. Montgomery and W. L. Brown, "Field-induced conduc-tivity changes in germanium," *Phys. Rev.*, vol. 103, pp. 865–870; August, 1956.

J. Bardeen, R. E. Coovert, S. R. Morrison, J. R. Schrieffer, and R. Sun, "Surface conductance and the field effect on germanium," *Phys. Rev.*, vol. 104, pp. 47-51; October, 1956.
 ⁸ A. Many and D. Gerlich, "Distribution and cross-sections of fast

states on germanium surfaces in different gaseous ambients," Phys.

Rev., vol. 107, pp. 404–411; July, 1957.
S. Wang and G. Wallis. "Recombination centers and fast states on unstable germanium surfaces," *Phys. Rev.*, vol. 107, pp. 947–953;

August, 1957. ¹⁰ O. M. Stuetzer, "Junction fieldistors," PROC. IRE, vol. 40, pp.

1377-1381; November, 1952. ¹¹ E. N. Clarke, "Nature of the water-vapor-induced excess cur-rent on grown germanium *p-n* junctions," *Phys. Rev.*, vol. 99, pp. 1899–1900; September, 1955.
 ¹² S. Penman and W. L. Brown, "The field effect in silicon," *Phys.*

 Rev., vol. 100, p. 1259; November, 1955.
 ¹³ G. G. E. Low, "Modulation of the surface conductance of germanium and silicon by external fields," *Proc. Phys. Soc. (London)*, vol. B68, pp. 10–16; January, 1955. ¹⁴ G. C. Dousmanis, "Semiconductor surface potential and surface

states from field-induced changes in surface recombination," ' Phys. Rev., vol. 112, pp. 369-380; October, 1958.

experiments using silicon as the base material. In general, though, the problem of obtaining an interpretable field effect on silicon has been rather formidable, and most experimenters have confined their work to the more responsive germanium.

EXPERIMENTAL TECHNIQUE AND RESULTS

This paper sets forth information obtained when dc electrostatic fields were applied normal to the surfaces of operating silicon *p-n-p* evaporative-fused-junction transistors. The primary parameter studied was α and the way it changed during the application of the field. For this study, apparatus was constructed in which a transistor could be placed in close proximity to a metallic field electrode. Fig. 1 is a schematic of the assembly pointing out the essential details. This method of assembly allowed us to achieve fields of up to 106 volts/cm before arcing ensued between the field electrode and the transistor. The resultant induced surface charge density produced by the maximum field was about 1.7×10^{12} charges/cm².

The capacitances in this assembly were small and did not appear to affect the field-effect measurements in any way. The capacitance of the field plate to the sample assembly was 9 $\mu\mu$ f.

The transistor was operated in the common base configuration with a 1-milliampere emitter current and a 5-volt collector bias. The small signal α was read manually and it usually took the operator from 5 to 20 seconds to read the α once the field voltage had been applied.15

A typical example of the effects observed in the experiments on α is illustrated in Fig. 2. The α of the transistor is plotted as a function of time. When a positive voltage was applied to the field plate [see Fig. 2(a)], α dropped initially and, while the voltage was still on, returned approximately to its original value as time passed. When the voltage was turned off, σ jumped to a new value, higher than the original one, and then decaved to its original value. Fig. 2 also shows the increase in α when a negative voltage was applied, and the subsequent decay in time. It can be seen from these results that turning off a positive voltage is analogous to turning on a negative voltage. The decay characteristics were reproducible to better than ± 15 per cent.

The observed decay in α can be correlated with the transfer of charge between the fast and slow states at the surface of the base region of the transistor.

¹⁵ Since these measurements were made, the equipment has been modified. A recording system which has a response time of 1 second is now in use.

Now

$$\mu = \frac{2\pi\Delta f}{2T_0}$$

where Δf = swept-frequency deviation. The outputinput peak power ratio is derived by squaring the amplitude of the output pulse, the input amplitude having been taken as unity. This vields

$$\frac{2\mu T_0^2}{\pi} = \left(\frac{4\pi\Delta f}{2T_0}\right) \left(\frac{T_0^2}{\pi}\right) = 2T_0\Delta f.$$
(25)

If the wide pulse width assumes the same dimensions as in the section on spectra derivation

$$2T_0 = T,$$

then the output-input pulse width and peak-power ratios become $T\Delta f$, if the convention is adopted that the output pulse is measured at the points $t = \pm \frac{1}{2}\Delta f$.

Acknowledgment

The investigation reported herein is part of the general program of the Sperry Gyroscope Co., for studying advanced techniques. W. W. Mieher and C. E. Brockner were chiefly responsible for the guidance and support necessary to provide program continuity. The author is also indebted to J. E. Chin, L. R. Sadler, and J. Cerar, who have made substantial contributions to the progress of the project.

CORRECTION

In a correction to the paper, "The Parametron, a Digital Computing Element which Utilizes Parametric Oscillation," by E. Goto, which appeared on page 1840 of the November, 1959 issue of PROCEEDINGS, R. G. Allen and J. E. Mezei have advised the editors of two typographical errors in the explanatory equations. For clarity, the entire correction is repeated here.

The five input parity-check circuit (Goto's Fig. 13) which should give a "1" output when an odd number of inputs are "1" does not appear to be correct, possibly because of an error in drafting.

Fig. 13 represents the following logical function:

$$f(x, y, z, u, v)$$

$$= \left[\left[x \ \bar{y} \ z \ \bar{u} \ v \right] \left[\bar{x} \ y \ \bar{z} \ u \ v \right] \left[x \ \bar{y} \ z \ \bar{u} \ v \right] \left[\bar{x} \ y \ \bar{z} \ u \ v \right] \left[\bar{v} \right] \right]$$

where the square brackets represent the majority function.

This five majority function can be reduced to the simpler function:

ĺ	A	В	L	E	Ī	

X	Y	5	26	T.	Goto Circuit Result	Desired Result
1	0	1	0	0	1	0
0	1	0	1	0	1	0
1	0	1	0	1	0	1
0	1	0	1	1	0	1

$$f(x, y, z, u, v) = \left[\left[x \ \bar{y} \ z \ \bar{u} \ v \right] \left[\bar{x} \ y \ \bar{z} \ u \ v \right] \left[\bar{v} \right] \right]$$

which does not vield the desired result in four cases (Table I) of the thirty-two possible combinations of five binary variables. A correct logical function for a five input parity check is

f(x, y, z, u, v)

$$= \left[\left[x \ \hat{y} z \ \tilde{u} v \right] \left[\bar{x} \ y \ \bar{z} \ u v \right] \left[\bar{x} \ y \ z \ \tilde{u} v \right] \left[x \ \hat{y} \ \hat{z} \ u v \right] \left[\bar{v} \right] \right]$$

which would be represented by the logical circuit shown in Fig. 1.



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Fig. 8—Compression-filter 1F waveform (lower half of waveform masked to reveal greater sidelobe detail).



Fig. 9—Multiple inputs and detected outputs of a pulse compression filter illustrating principle of superposition.

The pulse spectrum is

$$F(\omega) = \int_{-\infty}^{\infty} f(l) e^{-j\omega l} dl$$

=
$$\int_{-T_0}^{T_0} \exp\left\{j\left[(\omega_c - \omega)l + \frac{1}{2}\mu l^2\right]\right\} dl.$$
(17)

The generalized compression-filter function is

$$II(\omega) = \exp\left[j \frac{(\omega_r - \omega)^2}{2\mu}\right]$$
(18)

and the filter-output spectrum $G(\omega)=F(\omega)H(\omega)$

$$\therefore G(\omega) = \exp\left[j \frac{(\omega_r - \omega)^2}{2\mu}\right]$$
$$\int_{-T_0}^{T_0} \cdot \exp\left[j(\omega_r - \omega)l + \frac{1}{2}\mu l^2\right] dl, \tag{19}$$

The term g(t) represents the output time function where

$$g(l) = \frac{1}{2\pi} \int_{-\infty}^{\infty} G(\omega) e^{j\omega t} d\omega.$$
 (20)

Thus,

$$g(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left(\exp\left[j \frac{(\omega_c - \omega)^2}{2\mu}\right] \right)$$
$$\int_{-T_c}^{T_c} \exp\left[j(\omega_c - \omega)\tau + j \frac{\mu}{2}\tau^2\right] d\tau e^{j\omega t} d\omega. \quad (21)$$

This rearranges to

$$g(l) = \frac{1}{2\pi} \int_{-T_0}^{T_0} \left[\exp j \left\{ \frac{1}{2} \mu \tau^2 + \omega_r \tau + \frac{\omega_r^2}{2\mu} - (\omega_r + \mu \tau - \mu/)^2 / 2\mu \right\} \right] \\ - \left[\int_{-\infty}^{\infty} \exp \left[j (\frac{1}{2}\mu) \right] \omega - (\omega_r + \mu \tau - \mu \tau) \left\{ \frac{2}{2} \right] d\omega d\tau \quad (22)$$

letting

$$u = \frac{\omega - (\omega_{c} + \mu\tau - \mu t)}{\sqrt{2\mu}}$$
$$g(t) = \frac{\sqrt{2\mu}}{2\pi} \int_{-T_{0}}^{T_{0}} \left[\exp j \left\{ \frac{\mu\tau^{2}}{2} + \omega_{c}\tau + \frac{\omega_{c}^{2}}{2\mu} - (\omega_{c} + \mu\tau - \mu t)^{2}/2\mu \right\} \right] \left[\int_{-\infty}^{\infty} \exp (ju^{2}) du \right] d\tau, \quad (23)$$

but

$$\int_{-\infty}^{\infty} e^{ju^2} du = \int_{-\infty}^{\infty} (\cos u^2 + j \sin u^2) du = \sqrt{\pi} e^{j\pi/4}$$

and

$$g(l) = \sqrt{\frac{\mu}{2\pi}} e^{j(\omega_{c}t-1-2\mu t^{2}+\pi/4)} \int_{-T_{0}}^{T_{0}} e^{j\mu t\tau} d\tau$$

$$= \sqrt{\frac{2\mu}{\pi}} \exp\left[j\left(\omega_{c}l - \frac{1}{2}\mu t^{2} + \frac{\pi}{4}\right)\right] \int_{0}^{T_{0}} \cos \mu t \tau d\tau$$

$$= \sqrt{\frac{2\mu T_{0}^{2}}{\pi}} \frac{\sin \mu t T_{0}}{\mu t T_{0}}$$

$$\cdot \exp\left[j\left(\omega_{c}l - \frac{1}{2}\mu t^{2} + \frac{\pi}{4}\right)\right]. \qquad (24)$$

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World Radio History

pression performed by B. L. Hulland of the Sperry Gyroscope Co.)

The foregoing analysis showed that the compression filter would require a square-law phase characteristic of the form

$$\beta_f = - (\omega_c - \omega)^2/2\mu;$$

the associated time delay is then

$$t_d = \frac{d\beta_f}{d\omega} = (\omega_c - \omega)/\mu$$

This characteristic is physically unrealizable since it vields negative time delays over half the frequency band. This objection is met by adding a sufficiently large constant delay so that the filter time-delay is positive at all frequencies of interest. The modified filter timedelay is then

$$t_d' = (\omega_c - \omega)/\mu + k$$

and the phase characteristic to be approximated by the filter design is

$$\beta_{f}' = - (\omega_{c} - \omega)^{2}/2\mu + k\omega + C_{2}.$$

The type of network chosen for the filter design was the bridged-T equivalent (see Fig. 6) of the all-pass constant-resistance lattice network. Design techniques for all-pass networks are well covered in the literature,⁸⁻¹⁰ and the minute design details of a particular application need not be examined here. As the name implies, all-pass networks theoretically have no losses. Network purists will insist, and rightly so, that this characteristic cannot be obtained in practice, especially at frequencies in the IF range. However, judicious use of phase and amplitude compensating networks will provide the necessary engineering approximation for a system application.

Fig. 7 indicates a possible laboratory arrangement to test the pulse compression characteristics of a particular filter design. Increased compression ratios may be obtained by cascading additional sections of the designed filter. Figs. 8 and 9 indicate the type of test results that may be expected. Fig. 8 shows a representative compressed pulse, this being an IF waveform in which the lower half has been masked out to reveal the signal sidelobe detail more clearly. Fig. 9 shows the uncompressed pulses from two inputs and their respective video detected outputs from the compression filter for several degrees of overlap of the input signals. This demonstrates the linear operation of this technique in

⁸ O. J. Zobel, "Distortion correction in electrical networks with constant resistance recurrent networks," *Bell Sys. Tech. J.*, vol. 7, pp. 438–534; July, 1928.

⁹ E. A. Guillemin, "Communications Networks," John Wiley and Sons, Inc., New York, N. Y., vol. 2; 1935.
 ¹⁰ J. C. Pinson, "Transient Correction by Means of All-Pass Networks," Ph.D. dissertation, Mass. Inst. Tech., Cambridge; June, 1077.

1957.



Fig. 6-General form of bridged-T all-pass network.



Fig. 7-Compression-filter laboratory test equipment.

that the output response for broad overlapping pulses from multiple inputs will be precisely the superposition of the responses from each input when the others are absent. The amplitude ripples on the wide pulses illustrate the effect of the compression-filter delay line on the frequency-swept input signals.

CONCLUSION

A technique for increasing the average power capability in peak power limited pulse-transmission systems has been analyzed. The theoretical aspects are closely allied to matched-filter and cross-correlation methods. The derived pulse shape for this technique is $(\sin x)/x$, but the waveform may be modified by approaches that are analogous to the reduction of antenna-pattern sidelobes.

Appendix

The following closed-form solution of linear FM rectangular-envelope pulse compression is the result of the contributions of J. E. Chin to this study program,¹¹

$$f(t) = \exp\left[j(\omega_{c}t + \frac{1}{2}\mu t^{2})\right] - T_{0} \le t \le T_{0}$$

$$f(t) = 0 \qquad |t| > T_{0}. \quad (16)$$

¹¹ J. E. Chin and C. E. Cook, "The mathematics of pulse compression-a problem in systems analysis," Sperry Engrg. Rev. vol. 12, pp. 11-16; October, 1959.

and the spectrum phase-function is

$$\beta_{s} = (\omega_{c} - \omega)^{2}/2\mu - \tan^{-1} \left[\frac{S(x_{1}) + S(x_{2})}{C(x_{1}) + C(x_{2})} \right]$$
(12b)

where

$$x_1 = rac{\mu T}{2} + (\omega_c - \omega)$$
 and $x_2 = rac{\mu T}{2} - (\omega_c - \omega)$.

If, as implied in the previous section, matched filtering is attempted, the filter transfer characteristics $e^{-s/-j\beta f}$ must be conjugate to the spectrum function derived above, that is

 $e^{-\alpha_S} = e^{-\alpha}$

and

$$\beta_S = -\beta_f.$$

In practice, the phase characteristic of the compression filter is made to match only the imaginary square-law spectrum component, assuming that the residual phase term

$$-\tan^{-1}\left[\frac{S(x_1) + S(x_2)}{C(x_1) + C(x_2)}\right]$$

will not prove harmful. The consequences of this assumption are shown in the results of the Appendix.

The Fresnel functions do not represent a closed-form solution, and the spectrum functions must be derived from tables of Fresnel integrals.^{6,7} The Fresnel function argument is

$$y = \begin{bmatrix} \frac{\mu T}{2} \pm (\omega_r - \omega) \\ - \frac{\sqrt{\pi \mu}}{\sqrt{\pi \mu}} \end{bmatrix}.$$
 (13)

By making the substitutions

 $\mu = \Delta \omega / T$, $\Delta \omega =$ frequency deviation within wide pulse,

$$\Delta \omega = 2\pi/\tau$$
, $\tau =$ narrow pulse width $-\omega = n\Delta\omega/2$,

 ω_c then

$$y = \sqrt{\frac{\bar{T}}{\tau}} \left(\frac{1 \pm n}{\sqrt{2}}\right). \tag{14}$$

The argument, y, appears as a function of the compression ratio T/τ , and is seen to be independent of the absolute amount of the frequency deviation, $\Delta \omega$.

Fig. 5 illustrates the shape of the spectrum components for various values of compression ratio. As



various compression ratios.

this ratio increases, the amplitude distribution becomes more nearly rectangular and the residual phase component flat over the band of frequencies of major interest. The derivation of pulse shapes for these spectra made use of numerical summation. In each instance, the combination of the real and imaginary spectrum components produced a wave-form arbitrarily close to the $(\sin x)/x$ function and of the prescribed peak amplitude. As this study program developed, a closed-form solution for this method of pulse compression (linear frequency sweep within a rectangular pulse and a square-law phase compression filter) was obtained by J. E. Chin of the Sperry Gyroscope Co., and is given in the Appendix. This analysis showed that the filter-output pulse envelope is precisely of the $(\sin x)/x$ form. Moreover, it was shown that the carrier of the filter output pulse is frequency swept at the same rate as the input pulse but in the opposite direction. However, the total frequency deviation between the first zeros of the compressed pulse is given by

$$2\frac{\Delta\omega\tau}{T} \cdot$$
(15)

Whether this represents a serious problem or not will depend on the particular application and the compression ratio involved. Physically, the presence of this residual frequency modulation may be explained as arising from the frequency components introduced by the rise and fall portions of the wide-pulse envelope. These frequency components do not occur at the times dictated by the linear sweep of the carrier frequency. The resultant effect of the compression filter is to disperse these components in a manner to produce the reverse frequency sweep cited above. (This explanation results from the additional analysis of linear FM pulse-com-

⁶ E. Jahnke, and F. Emde, "Table of Functions," Dover Publications, Inc., New York, N. Y.; 1945. ⁷ A. Van Wijngaarden and W. L. Scheen, "Tables of Fresnel

⁷ A. Van Wijngaarden and W. L. Scheen, "Tables of Fresnel Integrals," Computation Dept. of the Mathematical Center, Amsterdam, The Netherlands, Rept. No. R49; 1949.



Fig. 3-Compression filter time-delay and phase shift.



Fig. 4-Wide-pulse and compressed-pulse waveforms derived by heuristic analysis.

yield an output pulse of width $\tau = 1/\Delta f$ (measured at the appropriate point) having an increased peak power ratio

$$\frac{\hat{P}_0}{\hat{P}_i} = \frac{T}{\tau} = T\Delta f.$$

The less obvious result was that the shape of the compressed pulse would be $(\sin x)/x$ and not rectangular. This may not be the most desirable pulse waveform for some radar applications because of the high sidelobe levels. Reducing these unwanted signals is entirely analogous to antenna-pattern sidelobe reduction, and some useful efforts in this area have been reported.4,5

DERIVATION OF LINEAR FM PULSE **COMPRESSION SPECTRA**

For the system under study, the transmitted signal function is

⁴ T. T. Taylor, "Design of Line Sources for Narrow Beamwidth and Low Side Lobes," Hughes Aircraft Co., Culver City, Calif., Tech. Memo No. 316; July, 1953. ⁶ C. E. Cook, "Modification of Pulse Compression Waveforms," presented at Next Electrowics Conf. Chicago III (October 15, 1059)

presented at Natl. Electronics Conf., Chicago, Ill.; October 15, 1958.

$$f(t) = A \cos(\omega_c t + \frac{1}{2}\mu t^2), \quad -\frac{T}{2} < t < \frac{T}{2}, \qquad (7)$$

where the carrier frequency is

$$\omega = \omega_c + \mu t.$$

If the constant A is neglected, the spectrum of this signal is

$$F(\omega) = \int_{-T/2}^{T/2} \cos(\omega_{c}t + 1/2\mu t^{2})e^{-j\omega t}dt$$

= $\frac{1}{2} \left[\int_{-T/2}^{T/2} \exp j \left[(\omega_{c} - \omega)t + \frac{1}{2}\mu t^{2} \right] dt + \int_{-T/2}^{T/2} \exp - j \left[(\omega_{r} + \omega)t + \frac{1}{2}\mu t^{2} \right] dt \right].$ (8)

The second integral essentially defines the spectrum at negative frequencies and has a negligible contribution at positive frequencies, provided the ratio $f_c/\Delta f$ is sufficiently large, which would be the case in any practical application of pulse compression.

The spectrum expression, after a suitable change of variables, becomes

$$F(\omega) = \frac{1}{2} \sqrt{\frac{\pi}{\mu}} e^{-j(\omega_{c}-\omega)^{2}/2\mu} \int_{\sqrt{\pi/\mu}}^{\sqrt{\pi/\mu}(T/2+(\omega_{c}-\omega)/\mu)} e^{j(\pi/2)x^{2}} dx$$
(9)

The above integral yields

$$F(\omega) = \frac{1}{2} \sqrt{\frac{\pi}{\mu}} e^{-j(\omega_c - \omega)^2 2\mu} \\ \cdot \left[C \left(\frac{\mu T}{2} + (\omega_c - \omega) - \frac{1}{\sqrt{\pi\mu}} \right) + jS \left(\frac{\mu T}{2} + (\omega_c - \omega) - \frac{1}{\sqrt{\pi\mu}} \right) \right] \\ + C \left(\frac{\mu T}{2} - (\omega_c - \omega) - \frac{1}{\sqrt{\pi\mu}} \right) + jS \left(\frac{\mu T}{2} - (\omega_c - \omega) - \frac{1}{\sqrt{\pi\mu}} \right) \right] (10)$$

where

$$C(x) = \int_0^x \cos\frac{\pi}{2} y^2 dy \qquad (11a)$$

and

$$S(x) = \int_0^x \sin \frac{\pi}{2} y^2 dy$$
 (11b)

are the Fresnel integrals.

Expressing the spectrum function in the form

$$F(\omega) = e^{-\alpha_S - j\beta_S},$$

then the spectrum amplitude-function is

$$e^{-\alpha_{S}} = \frac{1}{2} \left(\frac{\pi}{\mu} \right)^{1/2} \{ [C(x_{1}) + C(x_{2})]^{2} + [S(x_{1}) + S(x_{2})]^{2} \}^{1/2}$$
(12a)



Fig. 1—Idealized pulse compression characteristics. (a) Wide-pulse envelope. (b) Carrier-frequency modulation. (c) Filter time-delay characteristic. (d) Compressed-pulse envelope. (e) Input-output wave forms of compression filter.

supplied in the next section and in the Appendix.

The transmitted pulse is to have a rectangular envelope and a carrier frequency that is of the form

$$\omega = \omega_c + \mu t \qquad |t| < \frac{T}{2}$$
 (2)

The phase angle of the transmitted frequency becomes, when envelope contributions are ignored,

$$\phi = \int \omega dt = \omega_c t + \frac{1}{2} \mu t^2 + C_1.$$
 (3)

Thus, the phase angle ϕ is seen to contain a squarelaw term

$$\frac{1}{2}\mu t^2$$
. (4)

Further, if the product of the transmitted pulse width T and the frequency deviation $\Delta f = \Delta \omega/2 = f_2 - f_1$ is large, the linear progression of the carrier frequency between f_2 and f_1 should result in an essentially rectangular spectrum-amplitude distribution. Fig. 2 plots the essential features of the pulse derived by this method of reasoning.

The compression filter is to have a linear time-delay vs frequency characteristic of opposite sense to the linear frequency sweep. Functionally, this may be expressed as

$$d_d = 2K(\omega - \omega_1) + b.$$
 (5)

Since the filter is being used in a band-pass applica-



Fig. 2—Wide-pulse waveform parameters and assumed amplitude and phase spectra.

tion, the associated filter phase shift is

$$\beta_f = \int t_d d\omega = K(\omega - \omega_1)^2 + b\omega + C_2.$$
 (6)

It must be realized that in a practical filter design only that portion of the phase function which corresponds to a positive time delay can be synthesized. The relationships of (5) and (6) are plotted in Fig. 3.

If the constants μ and k are properly matched, the spectrum at the compression-filter output is assumed to consist of a rectangular amplitude distribution and a flat or linear phase component. The time function of the compressed pulse is easily recognized from the spectrum parameters as having a $(\sin x)/x$ envelope, the pulse width being $\tau = 1/\Delta f$ when measured 4 db down from the peak amplitude. The spacing between the first zeros of this envelope is $2/\Delta f$. The carrier frequency under the above assumptions is a constant, f_c , and the peak amplitude is, of course, $\sqrt{T\Delta f}$ (Fig. 4).

No loss of generality will result if the compression filter is assumed to have a bandwidth Δf and a rectangular amplitude response centered at f_c . This reduces the operation to matched filtering in the North sense if the second-order effects are ignored in this type of analysis.

The results obtained above confirm the earlier assumption that such a pulse compression system would

Pulse Compression—Key to More Efficient Radar Transmission*

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Summary-Increased demand for greater detection ranges in radar systems is often thwarted by the transmitting tube peak power limitation which, for narrow pulse operation, is usually reached before the full average power capability of the tube is realized. The technique of pulse compression offers a means of increasing the average power available to illuminate radar targets without any loss at the receiver of the resolution needed for the tactical requirements of the system. This is accomplished by transmitting a wide pulse in which the carrier is frequency modulated and then, by proper signal processing methods, causing a time compression of the received signal to a much narrower pulse of high effective peak power. The spectra and time functions of a particular class of pulse compression signals are analyzed and the basis for compression filter design is derived. Test waveforms demonstrate the resolving capability of the pulse compression technique.

INTRODUCTION

NINCE THE inception of the military applications of radar techniques, emphasis has been placed on extending the ranges at which objects may be detected. In most instances, the demand for increased detection range has not been at the expense of normal tactical requirements for a certain minimum amount of range resolving capability. Faced with this situation, radar tube designers have been forced to concentrate on stepping up the peak powers of their tubes, since the tactical considerations have not permitted extending detection ranges by increasing average power by means of a wider transmitted pulse. As a consequence, in many situations high-powered tubes are being used inefficiently as far as average power is concerned. To compensate for this inefficiency, engineers have developed post-detection integration techniques to extend the radar detection range. These techniques also lead to further inefficiencies as far as the use of total available average power is taken into consideration. It will be the purpose of this paper to study a technique for increasing the average power capability of a pulse radar so that there is neither an increase in peak power nor a degradation of pulse resolution.

PULSE COMPRESSION EVOLUTION

Several individuals have been concerned with the problem outlined above and have sought means for solving the problem of increasing radar detection range when the pulse width must be kept fixed and peak power limitations control the average power that may be used.¹⁻³ R. H. Dicke and S. Darlington in the United States have proposed more or less identical approaches. but on the basis of patent applications Dicke would appear to have priority of conception as far as the ideas discussed here are involved.

Dicke reasoned that if the carrier frequency of a transmitted pulse were linearly swept, as shown in Fig. 1(b), a pulse compression filter with the time-delay vs frequency characteristic of Fig. 1(c) could be used to delay one end of the pulse relative to the other. This would produce, at the filter output, a narrower pulse [Fig. 1(d)] which would be of greater peak amplitude. The linear time-delay characteristic of the filter would act to delay the high-frequency components at the start of the input pulse more than the low-frequency components at the end of the pulse, with frequency components in between experiencing a proportional delay. The net result would be a time compression of the pulse. Since a passive linear filter is postulated, the principle of the conservation of energy applies and the buildup in peak power of the compressed pulse would be proportional to the ratio of the widths of the filter input and output pulses. Thus

$$\frac{\vec{P}_0}{\vec{P}_i} = \frac{T}{\tau} \tag{1}$$

where

 $\hat{P}_i = \text{peak power input pulse},$ $\hat{P}_0 = \text{peak power compressed pulse}.$

If the pulse width τ represents the desired resolution, it can be seen that if this technique is feasible a pulse of width T, representing an increase in average power, may be transmitted with an associated frequency modulation that contains the information necessary to construct the desired compressed pulse of greater effective peak power. However, the actual peak power limitations of a pulse radar system are by-passed, thus opening another avenue for extending radar performance.

HEURISTIC ANALYSIS OF LINEAR FM Pulse Compression

The basis for undertaking investigation of the type of system postulated by Dicke stemmed from the heuristic reasoning given below. The more rigorous analysis is

^{*} Original manuscript received by the IRE, July 7, 1958; revised manuscript received May 28, 1959 and November 23, 1959. † Air Armament Division, Sperry Gyroscope Co., Great Neck, N. Y.

¹ R. H. Dicke, "Object Detection Systems," U. S. Patent No. 2,624,876; January 6, 1953.

² S. Darlington, "Pulse Transmission," U.S. Patent No. 2,678,997; May 18, 1954. ⁹ W. Cauer, German Patent No. 892,772; December 19, 1950.

CONCLUSIONS

The simulation experiments described above have gone a long way toward demonstrating the feasibility of a perceptron as a pattern-recognizing device. Both forced learning and spontaneous learning performances have been investigated, and some insight has been gained into conditions under which different systems break down, or deviate from typical biological learning phenomena. Although digital simulation is apt to be time-consuming and expensive, particularly for large networks, improved programming methods have cut down the running time considerably, so that for early investigations of all systems proposed up to this time, digital simulation is still competitive with the construction of actual hardware models. As the number of connections in the network increases, however, the burden on a conventional digital computer soon becomes excessive, and it is anticipated that some of the models now under consideration [8] may require actual construction before their capabilities can be fully explored.

Digital programs undertaken to date have been concerned exclusively with the logical properties of the network, rather than with any particular hardware embodiment; that is, there has been no attempt to introduce simulation of electronic noise, component variation, or other factors which might affect the performance of an actual system. The results of these programs, therefore, should be interpreted as indicating performances which might be expected from an "ideal," or perfectly functioning system, and not necessarily as representative of any particular engineering design. A Mark I perceptron, recently completed at the Cornell Aeronautical Laboratory, is expected to provide data on the performance of an actual physical system, which should be useful for comparative study.

A new program is currently being employed to simulate the "cross-coupled perceptron" described elsewhere [8]. The results of this study will be reported separately when they are available.

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response during the learning procedure, reinforcement being applied for whichever response is elicited by a given stimulus. The perceptron here was assumed to have an infinite number of A units, and the calculations were done with the third program, which was specifically designed to handle these conditions. The family of curves in Fig. 10(a) shows the performance as a function of the decay rate, δ . We find that for a zero decay rate, the system eventually learns to dichotomize the bars correctly 100 per cent of the time, *i.e.*, it learns to assign one response to all horizontal bars, and the opposite response to all vertical bars. However, this takes upwards of 3000 stimuli in most cases.⁴ As the decay rate increases, performance improves progressively, until a decay rate is reached (0.05 in this case) for which the system is unstable, and never attains perfect performance. The effect of the decay short of the instability level appears to be to keep previous reinforcements from accumulating to such a degree that they are difficult or impossible to undo, as the system settles into a more satisfactory terminal state; in other words, the decay keeps the system flexible, by making it possible to reverse the effects of previous learning more readily. At the instability level, previous reinforcements are reversed so readily that they are unable to maintain their effect at all, and associations are likely to be lost and reformed continually. The curve in Fig. 10(b) which shows expected waiting time to perfect performance, for the same series of runs, indicates the same phenomenon. We find that there is a clear optimum in performance as a function of the decay rate, for $\delta = approximately$ 0.01. Beyond this point, instability begins to occur, as indicated by the broken curve in the figure. This experiment is the best demonstration to date of

the "self-organizing" capability of a perceptron. Nonetheless, it can be demonstrated that minor changes in the stimulus environment will make it impossible for the same perceptron to achieve a satisfactory dichotomy. For example, if the 4 by 20 horizontal bars are replaced by double bars, composed of two 2 by 20 vertical bars separated by a space of 3 units, the perceptron will never spontaneously learn to distinguish the double bars from the single bars. Other classes of stimuli can be set up which are equally difficult, or impossible, for the system to learn spontaneously, although in each of these cases the problem would present no difficulty in a forced learning situation. Moreover, the curves in Fig. 10 are convex, indicating increasing difficulty in correctly associating the last few stimuli after most of the class has been learned. In a human subject faced with this task we would expect concave curves instead. These considerations indicate that the spontaneous learning capability of this perceptron, while interesting, is not sufficient to provide a basis for a biological theory of perceptual organization. This problem is considered in further detail elsewhere [8].

^{*} Individual runs differ from one another due to differences in stimulus sequence, even though the perceptrons are infinite; the curves shown are means of ten different runs

ten exposures, we find that the system has learned the "E" perfectly, but always gives the wrong response to stimuli of the opposite class (the letter "X"). The perceptron was then shown ten X's, to which the opposite response was forced, and we find at time 20 that it has now learned to give the desired response to the X, but has almost completely forgotten the proper response to the letter E. The amplitudes of such oscillations are apt to be increased by a large decay rate for the values of the A units (which makes more recent reinforcement more effective than earlier experience), but in the experiment illustrated here the decay rate was zero. Note that in Experiment 4-16 [illustrated Fig. 8(b)] the mean learning curve, shown by the broken line, climbs towards a high probability level as experience with both stimuli increases. At the same time, the swings in performance become considerably less pronounced, as each series of ten stimuli represents a progressively diminishing portion of the total experience of the system. The important conclusion from this experiment is that discrimination learning is possible for a linear system, provided the stimuli are sufficiently constrained in location. The retinal field in this case was 20 by 20 units, and the centers of the stimuli were constrained to a 5 by 5 region in the center of the retina. In Experiment 4-14 [shown in Fig. 8(a)], where the stimuli were distributed more freely over the retina (with the centers in a 13 by 13 field), no learning was demonstrated even after 200 stimuli. As a methodological experiment, these results indicate the importance of making sure that the stimulus distribution employed does not include "location cues" which are sufficient to indicate which stimulus is present, if we wish to test the ability of the perceptron to discriminate pattern characteristics exclusive of location. This can be fully guaranteed, in general, only by a uniform stimulus distribution over the entire field, with the elimination of special boundary effects by assuming a closed space, or an infinite space, as with the Born-von Kármán boundary conditions referred to in the Introduction.

Experiment 4-36, shown in Fig. 9, was again carried out with the second simulation program, this time with a more conventional perceptron. The threshold of zero, employed here, is sufficient to make the system fundamentally nonlinear, by eliminating the output of A units in the presence of negative input signals. The experiment was designed to show the performance of the system in the presence of a high degree of randomness, or noise, in the initial values of the A units. The stimuli for this experiment were vertical and horizontal bars, 4 units in width and 20 units long. A 5 per cent decay rate was introduced for the values of the A units. Note that in spite of the high decay rate and high initial noise level, the system achieved perfect performance on both classes of stimuli after a total of only 50 stimuli. This should be compared with the performance of very large (or infinite) perceptrons, in a spontaneous learning experiment with the same types of stimuli, which is illustrated in Fig. 10.



Fig. 9—Experiment 4-36. Forced learning experiment with vertical and horizontal bars, 500 A units, $\delta = 0.05$, $\theta = 0$, x = 4, y = 4, and $V_0 = \text{between} + 500$ and -500. Centers in 5 by 5 field, in 20 by 20 retina.



Fig. 10—Experiment 5-4. (a) Spontaneous organization of infinite perceptron in environment of 4 by 20 vertical and horizontal bars. (b) Expected waiting time to perfect performance, as a function of decay rate (means of 10 runs).

In the experiment shown in Fig. 10, stimuli were placed with equal probability at any position in a 20 by 20 retinal field, with Born-von Kármán boundary conditions. The stimuli were 4 by 20 horizontal and vertical bars, as in the previous case. The perceptron used in this experiment is one in which the A units are reinforced for the response R = 1, but are left unaltered if the response R = 0 occurs. Unlike all of the previously illustrated experiments, this is a spontaneous learning experiment, in which no attempt is made to control the

1.0



Fig. 5—(a) Experiment 10. "E" vs "X." No rotation. Centers placed in 13 by 13 field, in 72 by 72 retina. $N_A = 100$, $\theta = 2$, x = 5, and y = 5. (b) Experiments 20, 21. "E" vs "X" with shifting plus rotation. $N_A = 1000$, $\theta = 4$, x = 10, and y = 0.



Fig. 6—Experiments 16, 17, Square-diamond discrimination, $N_A = 1000$, x = 10, y = 0, and $\theta = 4$. Centers placed in 13 by 13 field.

fect response record after 60 training stimuli (30 of each class). In this experiment, of course, rotation was eliminated to avoid confusion of squares and diamonds, and the figures were merely displaced in the same manner as the E's and X's in the preceding experiment.

Fig. 7 shows two experiments concerned with partwhole discrimination, which was discussed in the preceding section. In Experiment 18, illustrated in Fig. 7(a) a system with only excitatory connections to the A units was simulated. The stimulus is shifted at random in the central portion of the field, as before. In this case, the letter "E" was correctly learned, but the system was unable to learn to give the opposite response to the letter "F." In Experiment 22, shown in Fig. 7(b), we see that



Fig. 7—(a) Experiment 18. "E" vs "F." $N_A = 100$, x = 10, $\theta = 4$, and y = 0. Centers placed in 13 by 13 field. (b) Experiment 22. "E" vs "F." $N_A = 1000$, x = 5, $\theta = 3$, and y = 5.



Fig. 8—Linear system experiments ("E" vs "X"). (a) Experiment 4-14, 15. $N_A = 500$, x = 4, y = 4. Centers placed in 13 by 13 field. (b) Experiment 4-16. $N_A = 500$, x = 4, and y = 4. Centers placed in 5 by 5 field.

a system, in which half of the connections to the A units are inhibitory, is able to learn the correct response to both classes of stimuli, although the F response is considerably less consistent then the E response. Experiments 18 and 22 are, unfortunately, not fully comparable, as the perceptron in the second case was a thousandunit system, while in Experiment 18 only a hundred A units were used. The character of the curves in these experiments, however, is definitely not a function of the size of the systems, but rather of the stimulus relationships, as shown by supporting evidence from many other cases. These results are in closer agreement with the theoretical predictions referred to earlier.

The next experiment (Fig. 8) was performed with the second simulation program, and represents the learning which is possible with a purely linear model, if the stimuli are constrained to one region of the retinal field. In this experiment, instead of testing the perceptron after every twenty stimuli, as in previous experiments, it was tested after every ten stimuli, which yields the characteristic pattern of converging oscillations shown in the figure. The first ten stimuli were all E's, and after these
put connections, excitatory connections alone being insufficient.³

Related to the problem of size variation in the stimuli is the problem of frequency variation, *i.e.*, some kinds of stimuli being more frequent than others. The response assigned to the more frequent stimulus type will generally tend to dominate the response assigned to the less frequent type, unless the system is designed in such a way as to minimize interaction between different classes of stimuli. The extent of this frequency bias was one of the problems originally set for the simulation programs, but a systematic investigation has not yet been completed.

A different problem area concerns the performance of linear systems. At one stage of the perceptron program, we were particularly interested in systems in which no threshold at all was employed in the A units, the output simply being equal to αv (the algebraic product of the input signal and the stored value) rather than $\alpha^* v$, as in the model described above. The values were to be augmented by a quantity equal to α if R = 1, and diminished by α if R = 0. It can easily be shown that in such a system, if a stimulus pattern can occur with equal probability anywhere in the retinal space (and eliminating special boundary conditions, as in the toroidally connected model), the expected value of every A unit after a long series of stimulus exposures will be exactly zero. Such a system clearly would not learn at all, if stimuli were distributed uniformly in space. If the stimuli were not uniformly distributed, however, the values would tend to correlate with any bias existing in the input signals, and it was predicted that such a system should learn to discriminate. The second simulation program was originally set up to study linear systems of this type, both in forced learning and spontaneous learning experiments. The theory of such systems in spontaneous learning is considered elsewhere [7]. While linear systems have now been abandoned, a typical experiment will be considered presently, as it illustrates several points of interest.

The problem of spontaneous learning—the ability of a perceptron to form meaningful classifications of stimulus patterns without any assignment of "correct" responses by a human experimenter—has prompted an extensive series of experiments. The effect was originally demonstrated with the second simulation program, where two disjunct classes of stimuli were properly separated, in a number of experiments. More interesting results were obtained with the third program, which eventually pointed the way to the development of the "cross-coupled association system," which promises to yield substantially improved performance in a large variety of problems [8]. In studying these spontaneous learning effects, the first question was whether they could actually be obtained at all, and the second was how much experience would be required, a question for which no satisfactory theoretical answer had been found at the time the simulation experiments were undertaken. In this area, there has been particularly close feedback between simulation work and development of the theory, the simulation program frequently demonstrating the existence of special cases, involving particular parameters or particular stimulus forms, which had not been anticipated. More recent theoretical models owe a great deal to this period of empirical exploration.

RESULTS OF SIMULATION EXPERIMENTS

The first experiments which we shall consider are concerned with the discrimination of the letters "E" and "X" in a forced learning situation, and are illustrated in Fig. 5. The stimuli were constrained to a central portion of the field (as shown by the insert) partly to facilitate learning, and partly to prevent truncation at the boundaries, since the toroidal stimulus space was not used in this program. Fig. 5(a) shows the probability of correct generalization (P_g) as measured on a sample of 20 X's and 20 E's. The stimulus sequence consisted of ten X's followed by ten E's, followed by a test of performance; then ten more X's, ten more E's, and a second test, for a total of 100 training stimuli. The data points shown in the figure are means obtained from ten 100 A unit perceptrons, each of them having a different connection network, but exposed to the same sequence of stimuli. The curves in Fig. 5(b) show the performance of a larger (1000 A unit) perceptron, on a more difficult variation of the same problem. In the solid curve, we see the performance of the system for stimuli rotated by some integral number of degrees selected at random between 0 and 30 degrees. This rotation is combined with vertical and horizontal translations selected within the same limits as in the preceding case. For rotations up to 30 degrees, note that the system attains perfect performance after only ten stimuli of each type. The broken curve shows the performance of the same system for rotations up to 359 degrees, combined with translations as above. In this case, there is a definite decline in the perceptron's performance, although it has attained a P_g of better than 0.90 after 30 stimuli of each type.

The next experiment (Fig. 6) was designed to check the hypothesis that performance on outline figures should be better than on solid figures, since unlike figures represented by their contours would have a minimum intersection on the retina, while solid areas might still have a large intersection even though their shape was different. The figures used were squares (illustrated in the inset) and diamonds, which covered the same areas as the squares, rotated 45 degrees. As shown by the two curves, the outline figures did indeed yield a better performance than the solid figures, giving a per-

³ F. Rosenblatt, "The Perceptron: A Theory of Statistical Separability in Cognitive Systems," Cornell Aero. Lab., Buffalo, N. Y., Rept. no. VG-1196-G-1; January, 1958. See p. 53.

tional training has improved the performance, and thus continue alternating between training and testing programs indefinitely. It is also possible to reverse the assigned responses in the middle of the experiment, thus reversing previous learning. In order to obtain unambiguous comparisons of performance in different parts of the training series, the testing series are generally "primed" with the same random number to guarantee that the same stimulus transformations will be used on each repetition of the program. The training programs, on the other hand, continue to select stimuli at random, independently of what has gone before. A comparison of the organization of the training and testing programs is presented in the flow diagrams in Fig. 3.

The two main simulation programs total about 5000 words each. The first program was designed to handle up to 1000 A units, and a 72 by 72 sensory mosaic. It was found that this large sensory system presented stimuli with a fineness of grain considerably better than the limits of discrimination of a thousand-unit perceptron, and at the same time, required an excessive amount of time for stimulus transformations, since each illuminated point in the stimulus must be transformed individually into its image point. The second program reduced the retina to a 20 by 20 mosaic, and limited the number of A units to 500. For the first system, the computing time averaged about 15 seconds per stimulus cycle, while in the second system the time was cut to about 3 seconds per cycle. Subsequent improvements in programming techniques indicate that it should be possible to reduce the computing time still further—say to about one second per cycle-for perceptrons of the size allowed by the second program. At the same time, however, analytic developments have suggested a way of actually calculating the exact performance of a given perceptron of the type discussed above, provided all possible stimuli are known, and a matrix of g coefficients, describing the interactions of each pair of stimuli, is computed for the particular network in question. This technique is discussed in the appendix to [7], and is the method employed in the third of our simulation programs for the analysis of spontaneous learning in infinite perceptrons. In that program, the response of the system is obtained analytically, rather than simulated, but the sequence of stimuli is governed by a series of random numbers generated by the program. We will consider some of the results of this program later in this paper.

THEORETICAL PREDICTIONS AND PROBLEMS

Before considering the results of the simulation experiments, let us review the main predictions coming from the theory of the perceptron (see [5]-[7]). The simulation experiments were designed in part to verify these predictions, and in part to study problems which were suggested by the theoretical investigations.

Fig. 4 shows a set of theoretical performance curves for perceptrons of three different sizes, in the problem of discriminating a square from a circle. The broken curves (for P_r) show the probability of giving the correct response to a stimulus which is identical in position, size, etc., to one which was shown previously, during the training period. The horizontal axis indicates the number of stimuli of each class (squares and circles) which were presented during the training period. The solid curves indicate the probability of correct response to any square or circle, regardless of whether it was used as a training stimulus or not. Note that both sets of curves approach the same asymptotes as the number of training stimuli becomes large. The first task of the simulation program was to check the general character of these learning curves for typical stimulus material, such as letters of the alphabet or geometric patterns. In particular, it was essential to determine whether the rates of learning agreed with the predicted rates, at least to a reasonable approximation.



Fig. 4-Learning curves for three typical perceptrons.

A second problem concerned the effect of particular types of transformations, such as shifting of stimuli, rotations, or size changes, upon the learning curves. The original theory did not distinguish among these types of transformations, and it was important to find out whether the system would work equally well for all of them. While sufficient demonstrations have now been made of performance under shifting and rotation conditions, the problem of size changes remains a serious one. with a number of special cases. One such special case involves the assignment of different responses to two stimuli, one of which could be considered a "part" of the other, such as a small circle which could be completely imbedded in a larger one, or the letter "F," which can be considered as an "E" with the lower bar missing. It was predicted that such discriminations would be possible only with a mixture of excitatory and inhibitory in-

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the state of each sensory point is indicated when a "visual" pattern is presented.² The perceptron construction routine prepares a table listing all of these connections. In the first simulation program this table was stored on tape; but in the second program, by cutting down the admissible number of A units and connections, it was possible to store the entire table in the core memory, saving a factor of about five in running time of the program. The R units to which each A unit is connected are similarly assigned at random in each of the first two programs, which permit multiple output connections from each A unit. Since, in practice, all experiments have been concerned only with simple binary discrimination problems, more recent programs have been designed with only one R unit, to which all A units are connected. In the second program, it is also possible to assign an initial random distribution of values to the A units, although in most experiments it is assumed that the values start out uniformly from zero.

2) The second stage in the experiment calls for reading a set of "prototype stimulus patterns" into the memory of the computer. These patterns consist of actual dot images of the stimuli to be used, punched as patterns of holes in IBM cards. Thus, if it is planned to teach the perceptron the first four letters of the alphabet, we would read in the images of the letters A, B, C, and D, which are stored for future reference by later routines. These prototypes are never altered, but are used by the stimulus transformation routines which are included in the remaining two programs, to construct variously displaced, rotated, or contracted patterns which are the stimuli actually "shown" to the perceptron.

3) Having constructed the connection tables and read in the prototype stimuli, the computer is ready to begin the actual learning experiment. This consists of an alternation between the two remaining programs, one of which attempts to "teach" the perceptron to recognize the stimulus patterns, while the other evaluates the performance of the perceptron at intervals specified by the control cards. For example, in a typical experiment, the discrimination of the letters "E" and "X," the procedure is as follows. First, a control card calls for the training program to show ten different transformations of the letter "E" (the first stimulus). Each of these is generated by applying a vertical and lateral shift of random magnitudes between zero units of retinal distance and a maximum shift specified by the control card, a rotation between zero degrees and a specified maximum, and a size somewhere between a specified lower and upper bound. Random numbers generated by the routine determine the exact transformation to be applied to each stimulus, and a new image is composed. The control card then specifies that the response "1" is to be reinforced as the appropriate response for the letter "E." The program accordingly calculates the signals received by each A unit from the transformed stimulus, determines which A units are active, and reinforces the units according to the rules for reinforcement of the R = 1 condition, for the gamma system, *i.e.*, each active A unit gains an increment in value, while the inactive units lose a compensating amount. In the second of the simulation programs, it is also possible for the stimulus to persist for a designated number of cycles, undergoing a random walk during this time, consisting of unit displacements, rotations, or size changes from the position in which it first appeared. This procedure is characteristic of the "forced learning mode" of experiment, which is the only mode possible for the first simulation program. In this mode, the desired response is turned on, or forced, by the training program at the same time that a stimulus is presented. The second program is also designed to permit a "spontaneous learning mode," in which stimuli occur in a random sequence, and the response spontaneously occurring upon presentation of the stimulus is reinforced, regardless of whether or not it is the response ultimately desired. Most of the experiments to be described in this paper were performed in the forced learning mode. After having presented the ten transformations of the letter "E" which were called for, and reinforced the response R = 1 for each transformation, control is returned to the supervisory routine, which reads the next control card. In this typical experiment, we next call for ten transformations of the letter "X," to be associated to the response R = 0. This procedure is carried out in the same manner as before.

We now switch to the testing program, which composes a series of stimulus transformations in the same manner as the training program, and goes through an identical set of calculations to determine the active A units in each case. Instead of reinforcing the association units, however, this program merely records the response, and checks it against the desired response for correctness. If the response is correct, it increments a tally of correct responses. Typically, we may look at twenty transformations of the "E" and twenty transformations of the "X," determining in each case the percentage of correct responses (R = 1 or R = 0, respectively). During this procedure, a running description of the responses of the system, numbers of active units, and other analytic data, are printed out by the computer. We may now present another ten E's and another ten X's, reinforcing the system as before, then test the performance once more, to find out whether this addi-

² In each of the first two simulation programs, multiple connections from the same A unit to the same S point are prohibited. In the second program, an inverse constraint was originally employed, fixing the number of connections originating from each sensory point, and assigning termini at random in the association system. This was later modified by the addition of a scheme to obtain, as nearly as possible, uniform numbers of inputs to each A unit as well as fixed numbers of outputs from the sensory units. These variations have not seriously affected the performance of the program, but it appears that somewhat better performance is obtained with the numbers of inputs to the A units is kept uniform.

random to a large number of "motor area" cells, the cells of the association system are connected to one or more binary response units, which are turned to their "1" state if they receive a positive signal from the association system, or to their "0" state if they receive a negative signal. The magnitude of the output signal generated by an active A unit is called the "value" of that unit, and is represented by the symbol v. The values of the units are stochastic variables, which change as a function of the history of the system. The organization of a simple perceptron with a single binary response is shown in Fig. 2. The total signal delivered by the set of A units is equal to $\sum \alpha_i^* v_i$ where α_i^* is equal to 1 if unit a_i is active, and 0 if a_i is inactive, and v_i is the current value of unit a_i . Note that there are two feedback lines from the response unit (or R unit) to the set of A units. These feedbacks control the "reinforcement," or changes in value, of the A units. In general, if the response R=1 occurs, active A units will gain in value, while if the response R=0 occurs, active units will lose in value. The value of the A unit thus acts as the memory variable for the system. It has been shown to be desirable to further modify the values of the A units by the rule that if some subset of units gains or loses in value, then the remainder of the units must change in the opposite direction just sufficiently to balance out the net change to zero. Thus, one unit can only gain parasitically, at the expense of the other units, and the total value of all of the A units is kept equal to zero at all times. A perceptron with this property is called a "gamma system." The theory of such systems has been considered in detail elsewhere [5], [6].



Fig. 2-Organization of a simple perceptron.

DESCRIPTION OF SIMULATION PROGRAMS

Fig. 3 shows the organization of a typical simulation program, for the study of perceptron performance in an environment of visual forms. Actually, four basically different programs have so far been written with a number of variations of each, but the two programs which were used for most of the experiments reported here are both organized in the manner illustrated. The third program involves more direct methods of computation rather than true simulation, while the fourth program (designed to study "cross-coupled systems," in which A units may be connected to one another as well as to S points and R units) has proven too slow to be used successfully.¹ The simulation programs have four main tasks, each of which is actually performed by a separate, self-sufficient program, which is stored on tape, and called into the computer by a supervisory routine. The supervisory routine reads instruction cards provided by the experimenter, which provide information on parameters, and control the sequence of subprograms performed in the course of the experiment. When each subprogram has been completed, control is passed back to the supervisory routine, which reads the next card for further instructions. In a typical experiment, the sequence is as follows:



Fig. 3—Flow diagram for simulation program.

1) The perceptron construction routine is called into the core memory, and reads in a set of parameters describing the perceptron to be constructed. These parameters include the number of A units, the number of excitatory and inhibitory connections to each unit, the thresholds of the units, the number of R units, the number of R units connected to each A unit, the decay rate for A unit values (which decay with time in some models) and a random number to be used for priming the pseudo-random-number generator used to control the choice of connections. The program then selects for each A unit a set of x + y sensory points to be assigned as origins for the input connections. This is done by generating a random number number modulo N_s (the number of sensory points) for each connection. This number is used to locate one of the N_s storage locations in which

¹ The cross-coupled system was successfully simulated, and predicted effects obtained in December, 1959, using an improved program. Results will be reported in later publications.

2) Suitable measures of performance must be defined. This means that some task must be set for the system, the outcome of which can be clearly recognized, and, preferably, counted or quantified in some manner. Signal strengths, waiting times for achievement of a criterion, or percentage of correct decisions are examples of suitable measures.

3) Experiments should be designed with suitable controls against trivial or ambiguous results. If we are interested in teaching a device to generalize a response to visual forms, for example, it is essential that a discrimination test should be made involving at least two different responses, to make sure that the system has not simply generalized the desired response universally to all stimuli, regardless of their similarity to one another. Moreover, it is often important to make sure that the cue for the response is the actual form of the stimulus, rather than its location on the retina, or some other unintentional source of information. This last condition is often quite tricky to satisfy, and in most of our current work we make use of Born-von Kármán boundary conditions (in which patterns shifted off of one edge of a retinal field re-enter on the opposite side, as in a toroidally connected space) in order to guarantee the logical equivalence of all points in the retinal space. Given such a retinal field, it is sufficient to place each stimulus pattern with equal probability or frequency at all possible locations in the retinal space, in order to guarantee that the illumination of a particular retinal point does not convey any information about which stimulus is present. It should be noted that this condition is not always observed in the experiments reported in this paper, stimuli often being confined to some subfield of the retina in order to increase the rate of learning. In at least one case (the experiment with the "continuous transducer perceptron" shown in Fig. 8) a discrimination has thus been obtained which would not hold up if the field were uniformly covered with the stimulus patterns.

ORGANIZATION OF A PERCEPTRON

Any perceptron, or nerve net, consists of a network of "cells," or signal generating units, and connections between them. The perceptron is defined by two sets of rules: 1) a set of rules specifying the topological constraints upon the network organization, such as the number of connections to a given unit, or the direction in which connections are made, and 2) a set of rules specifying the dynamic properties of the system, such as thresholds, signal strengths, and memory functions. A "fully random network" would be one in which only the number of cells and the number of connections is specified, each connection being equally likely to originate or terminate on any cell of the system. The topological rules for the organization of a perceptron take the form of constraints applied to such a random network, and it is assumed that all connection properties other than those specified remain "random," in the sense just indicated.

A simplified version of the known features of a mammalian visual system is shown in Fig. 1, for a comparison with the organization of a perceptron, which will be described presently. At the extreme left we see a mosaic of light-sensitive points, or retina, from which signals are transmitted to the visual projection area, in the cerebral cortex. Several intermediate relay stations exist in a typical biological system, which are not shown here. These connections preserve topological characteristics of the stimulus in a reasonably intact form. Beyond the projection area, however, connections appear to be largely random. Impulses are delivered through a large number of paths to the association areas of the cortex, where local feedback loops are activated, so that activity may persist for some time past the termination of the original visual stimulus. From the association area, signals are transmitted to the motor cortex, which again has a clear topological organization corresponding to the location of muscle groups to be controlled.



Fig. 1—Organization of a biological brain. (Heavy black areas indicate active cells, responding to the letter X.)

This general plan of organization has been considerably simplified in the perceptron. First of all, we will eliminate the projection area, and assume that the retinal points are directly coupled to association cells, or "A units." The number of input connections to each A unit is specified, but the locations of the origin points for the connections are selected at random from the set of sensory points. Each A unit receives some number, x, of excitatory connections, and some number, y, of inhibitory connections. The connection system from the sensory to association system is a many-to-many system. An excitatory connection from an illuminated sensory point is assumed to transmit a unit positive signal, while an inhibitory connection carries a unit negative signal. Each A unit has a fixed threshold, θ , and is triggered to deliver an output pulse if the algebraic sum (α) of the signals received from the x + y input connections is equal to or greater than θ . A further simplification is introduced at the output side of the association system. Instead of delivering its output signals at

Perceptron Simulation Experiments*

Summary—An experimental simulation program, which has been in progress at the Cornell Aeronautical Laboratory since 1957, is described. This program uses the IBM 704 computer to simulate perceptual learning, recognition, and spontaneous classification of visual stimuli in the perceptron, a theoretical brain model which has been described elsewhere. The paper includes a brief review of the organization of simple perceptrons, and theoretically predicted performance curves are compared with those obtained from the simulation programs, in several types of experiments, designed to study "forced" and "spontaneous" learning of pattern discriminations.

INTRODUCTION

NUMBER of papers and reports have been published describing the theory of a new brain model called the perceptron. The perceptron is a minimally constrained "nerve net" consisting of logically simplified neural elements, which has been shown to be capable of learning to discriminate and to recognize perceptual patterns [5]-[8]. This paper is concerned with a report of digital simulation experiments which have been carried out on the perceptron, using the IBM 704 computer at the Cornell Aeronautical Laboratory. These experiments are intended to demonstrate the performance of particular systems in typical environmental situations, free from any approximations which have been used in the previously published mathematical analyses. In the simulation programs, the action of every cell and every connection in the network is represented in detail, and visual stimuli are represented by dot patterns corresponding to illuminated points in a retinal mosaic.

Several related experiments have been conducted previously, using a digital computer for the simulation of a nerve net in learning experiments [1], [2], [4]. Rochester and associates, at IBM, have reported on several attempts to simulate the formation of "cell assemblies," in a model based on the work of Hebb [3] Hebb proposes that a set of neurons which is repeatedly activated by a particular sensory stimulus becomes organized into a functional unit, which can be triggered as a whole by sensory patterns sufficiently similar to the original one. Hebb's book, however, does not attempt to specify in a rigorous manner the exact organization or parameters under which the predicted effects would be obtained, so that the IBM group found it necessary to improvise several models and variations of their own, having various degrees of biological plausibility, in an attempt to construct a definite system. The results of these experiments seem ambiguous, not only because

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† Cornell Aeronautical Lab., Inc., Buffalo, N. Y.

of the uncertain relationship of the final model to the nerve net originally suggested, but also because the phenomenon which was sought after has never been defined in a fashion precise enough so that one might say whether or not it has actually occurred. These experiments illustrate the importance of selecting a suitable measure of performance in work of this type; it is essential that a clearly defined test should be specified for the "learning" which has presumably taken place, or else it is impossible to say either how well a particular system has performed or to compare its performance with any other system, or class of systems, in a systematic fashion.

From this standpoint, the experiments reported by Farley and Clark [1], [2] seem to have been better conceived. In this model, a network of eight randomly connected neurons was simulated. Inputs consisted of stimuli applied to one of two disjunct pairs of "input cells," and outputs were measured as the activity of two pairs of "output cells." In later experiments, the size of the network was increased to sixteen cells. It was demonstrated that this system can learn to favor the output from one set of output cells following the presentation of one of the two stimuli, and the alternative output set following presentation of the other stimulus. The problem of generalization was considered only in terms of relatively slight displacements or alterations of the stimulus patterns, and it was suggested that, under these conditions, the response would be most likely to occur which was previously associated to the stimulus having the greatest overlap with the altered stimulus. The problem of generalization to similar but completely disjunct stimuli was not specifically considered. Nonetheless, the process of generalization advocated as a result of these experiments has much in common with our early work on the perceptron. A more thorough consideration of this problem will be published elsewhere [8].

The design of a simulation program for studies of pattern recognition and perceptual generalization in nerve nets should fulfill at least three basic conditions, each of which has been ignored too frequently in previous work along these lines.

1) Simulation should not, in general, be attempted without a theoretical analysis of the nerve net in question, sufficient to indicate suitable parameters and rules of organization, and to indicate questions of theoretical interest. The examination of arbitrary networks in the hope that they will yield something interesting, or the simulation of networks which have been specially designed to compute a particular function by a definite algorithmic procedure seem to be about equally lacking in value. be necessary to remove it and introduce a new intact element. Such a removal could be accomplished by a lapping or grinding operation which either pulverized or cut out the element. Subsequently, a new element could be put in place and suitably connected. Again, this operation, once worked out and suitably tooled, does not require engineering labor. However, this method assumes elements of such a large size that they may be handled individually.

C. Repair of Integrated Devices

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FACTOR(F

COST

Although it is entirely possible to integrate into one single device, or package, the main part of a computer, for example, the extent to which integration should be carried would be limited by logistic considerations. Assume that one element of the computer started to malfunction after a certain time. Then it is desirable to repair the computer by exchanging a certain part, but usually not the entire computer. The manner in which the computer should be departmentalized to facilitate and minimize the cost of repair would have to be determined by logistic considerations. This requires knowledge about the mean time between failures, and therefore cannot yet be done with confidence. However, even from this point of view, a limit below 100 stages in one department of the device appears reasonable.

IV. OTHER FACTORS THAT INFLUENCE AN EVALUATION

Many other factors have been suggested that would affect an evaluation of integrated devices, such as cost of

> 20 30 SHRINKAGE S(PERCENT)

Fig. 12-Cost factor of a doctored 10-stage integrated device vs shrinkage. Shrinkage in doctoring equal to over-all shrinkage.

encapsulating a large unit compared to many small units, the utilization of large but faulty integrated devices for the fabrication of several smaller units, etc. It appears, however, that at the same time as they seem less fundamental than the shrinkage rate, they are also less amenable to a rigorous analysis and therefore will be omitted here. Also omitted will be factors that are characteristic of a specific device or a specific fabrication method.

While in theory it is possible to make almost any communication circuit into an integrated device, it is most easily realized in the case of iterated stages of identical construction, which give very simple operation such as the ON-OFF circuits of digital computers and automation equipment. Also the very large numbers of such circuits used make the gain in weight and space worthwhile. The higher cost of integrated devices, compared to assemblies of individual units, suggests the application to those in which there is a sufficient premium on weight and space. This is usually the case in mobile installations, particularly military installations, and it is here that integrated devices are most needed. However, in the long run it is believed that microminiature electronic systems have a more important function in bringing to a manageable size the complex equipment needed to relieve man from routine mental labor of all types.

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Fig. 13-Cost factor of a doctored 10-stage integrated device vs shrinkage. Shrinkage in doctoring negligible.



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COST FACTOR

B. Shrinkage Reduction by Reserve Stages

Several methods may be thought of to reduce the influence of shrinkage. One such method would be to improve the faulty stages by means such as selective etching to remove weak parts of a junction, sandblasting to increase surface recombination over a small area, replacing broken leads, etc. However, this requires diagnosis and choice of correct doctoring method, something that can be done only by skilled labor and which would therefore be costly. Another method, and one that is well-adapted to routine operation, is that of reserve stages. This method will be considered in this section.

As an illustration consider a shift register such as the one shown in Fig. 1. The yield of units with ten good stages, Y_{10} , the yield of units with 9 good stages, Y_{9} , etc., is obtained from

$$Y_n = (1 - S/100)^N (S/100)^{10-N} \frac{|10-n|}{|n|}, \qquad (8)$$

where n is the number of good stages. These yields for different values of shrinkage are shown in Fig. 11. For 0 per cent shrinkage, the yield of 10-stage units is, of course, 100 per cent, and for higher shrinkage, the yield is pushed continuously towards units with fewer and fewer stages. Let us assume that the completed register should have 10 good stages. Suppose that in designing the device two extra stages are added, making 12 in all.

Fig. 11—Yield of integrated units with 10, 9, 8, etc., good stages vs shrinkage in fabrication of individual stages,

After the devices are completed, but not encapsulated, they are tested and, in all the devices with one faulty stage, the base contacts of the adjacent stages are connected. This insures that when one of those stages goes ON, its base contact will turn on the other one in the pair simultaneously, thus bypassing the faulty stage. This scheme may be extended to still further stages so that with a 16-stage device, 3 faulty stages may be bypassed, etc. However, the addition of extra stages, of course, increases the shrinkage as shown in Fig. 7, and finally an optimum is reached.

The same method is applicable to other forms of integrated devices. In the case of the multiple AND gate, shown in Fig. 2, a metallic connection may be introduced from the source of one unit to the drain on the unit on the other side of the faulty stage, thus effectively bypassing the faulty stage.

Assume that the cost of bypassing a faulty stage is 10 per cent of that of making a new device. Further assume that the bypassing has no shrinkage. Then the cost factor becomes

$$F = \frac{(1 - S/100) \left[1 + \frac{1}{10} \sum_{1}^{m} \frac{Y_{m}}{100} \right]}{(1 - S/100)^{N} + \sum_{1}^{m} (1 - S/100)^{m} Y_{m}},$$
 (9)

where Y_m is the relative yield of units with *m* stages faulty and *N* is now the total number of stages:

$$N = 10 + 2m.$$

For the case in which the bypassing has the same shrinkage rate as the main manufacturing process, the cost factor becomes

$$F = \frac{(1 - S/100) \left[1 + \frac{1}{10} \sum_{1}^{m} \frac{Y_{m}}{100} \right]}{(1 - S/100)^{N} + \sum_{1}^{m} Y_{m}}$$
(10)

These expressions are shown graphically in Figs. 12 and 13. Although these assumptions may represent an oversimplification, and though other assumptions may alter the cost factor somewhat, the trend is clear. By providing reserve stages and using them to bypass faulty stages in the fabrication procedure, a considerable reduction in shrinkage can be accomplished. It appears that a design that facilitates doctoring of this type would be desirable in the fabrication of integrated devices.

While this method of bypassing is applicable to many integrated devices, it is not as useful in integrated devices involving matrices. Here the bypassing of one element may throw out not only the element but the complete row of which this element was a part, the column of which the element was a part, and also in some cases the corresponding rows and columns of matrices and whatever other devices would be combined with them. To limit the damage caused by a faulty element, it may



It seems therefore that an important requirement for integrated devices is a manufacturing process with a sufficiently low shrinkage. A somewhat higher investment in the fabrication equipment for integrated devices, compared to that for individual units, is justified because of the higher cost of shrinkage for the former.

This is illustrated in Fig. 9. Here the fabrication cost has been divided into two parts, one part consisting of investment in the fabrication process, which is independent of the number of units made. Such costs would be research and development, acquisition of machines, improvements in the process, automation, etc. The other part is processing cost which is directly proportional to the number of units made. Such costs are materials and labor. Of these two costs, F_1 varies with shrinkage as shown in Fig. 7. For F_2 let us make the simple assumption

$$S = 1/(F_2 + 1). \tag{7}$$

This means that S can be reduced arbitrarily by spending more on F_2 , as is found in practice. The boundary values are

$$S \to 0$$
 when $F_1 \to \infty$,
 $S \to 1$ when $F_2 \to 0$,

again in agreement with practice.

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Plotting F_2 along the abscissa and using the relations (5) and (7) gives the dashed lines in Fig. 9. The straight line for the investment cost has been given a slope corresponding to a certain number of units fabricated. The resultant total cost is shown in solid lines.

The total cost per stage for single devices shows a minimum which is the optimum point for fabrication of conventional devices. At this point, corresponding to a shrinkage of approximately 30 per cent and a fairly low investment, the cost of a five-stage integrated device is very large. Increasing the investment in the fabrication method, however, sharply reduces the total cost for the integrated device until a minimum, representing the optimum point for fabrication of integrated devices. is reached at a higher investment. Although this mathematically exact treatment represents an oversimplification of the practical situation, it is qualitatively correct.

Furthermore, it is desirable that the integrated device be designed so that the extent of integration—the number of components that are integrated into one piece of semiconductor—is flexible and adjusted to the particular shrinkage rate at any one time. An example is illustrated in Fig. 10.

For many semiconductor devices at the present time, the over-all shrinkage may well amount to 30 per cent. This figure, however, usually includes some correlated shrinkage and corresponds to a somewhat lower figure in Fig. 8, say approximately 15 per cent. For most commercial electronic equipment used at the present time, cost of a miniature version may not often exceed 5 times that of a conventional version. Then the maximum amount of integration should not exceed some ten stages. Shrinkage, therefore, represents, at the present time, the most stringent limitation on integrated device design.

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Fig. 10-Two integrated devices with identical cost factor F = 1.5. (a) $\tilde{S} = 10$ per cent. (b) S = 30 per cent.



A completely general treatment would assume an arbitrary correlation coefficient, c_J , where $0 < c_f < 1$. Here $c_f = 0$ corresponds to the first category and $c_f = 1$ to the second category mentioned above. However, lacking this general treatment, let us investigate the worst possible case corresponding to $c_f = 0$.

Let us assume that an integrated device offers such advantages that its cost may be larger than that of its corresponding assembly of individual units, by a factor F, where $F \ge 1$. Then the shrinkage of the integrated device may be allowed to reach, but should not exceed, $(F-1/F) \times 100$ per cent.

Consider now as an example, a shift register, and compare an *N*-stage device of this type with an assembly of *N* individual active devices. Let us disregard the passive components for a first approximation. Assume that the integrated device has a shrinkage such that *S* out of every 100 stages cannot be used. Then the yield of complete integrated devices, Y_{i} , is

$$V_i = (1 - S/100)^N \times 100 \text{ per cent.}$$
 (2)

For the individual units assembly, it is fair to assume the same shrinkage per unit. With no shrinkage in the assembling process, the yield of subassemblies, Y_{s} , is

$$Y_s = (1 - S/100) \times 100 \text{ per cent.}$$
 (3)



Fig. 7—Cost factor of integrated device vs shrinkage in fabrication of individual units.

Making up for the lower yield of integrated devices by F times higher cost gives

$$Y_i/Y_s = (1 - S/100)^{N-1} = 1/F,$$
 (4)

or

$$F = (1 - S/100)^{1-N}, \tag{5}$$

In Fig. 7 are shown some values of S and N of practical interest, and corresponding values of F are computed from (4). As an example, it can be seen from the curves that the cost of a 10-stage integrated unit is 7 times that of an assembly of ten individual units when the shrinkage on the fabrication process amounts to 20 per cent. However, for large shrinkage, the cost of any but the smallest number of stages probably is prohibitive for most applications.

Shrinkage is a phenomenon that is seldom constant but may vary up and down from some long-time average value. For a comparison, the long-term average value of F may be computed from a summing up of yields over the entire period,

$$1/F = \sum_{p} 1/F_{p}, \tag{6}$$

where F_p is the cost factor for a certain part of the period considered.

For a particular application, such as satellite instrumentation, the small size of an integrated device may allow a cost that is, for example, 1.5–5 times that of an assembly of individual units. The maximum number of stages that may then be economically used in each integrated device is shown in Fig. 8. This number decreases extremely rapidly with shrinkage.



Fig. 8—Maximum number of stages in an integrated device vs shrinkage. Cost-factor parameter.

devices to some reasonable number of stages. A size of the order of 100 stages seems within reason.

Heat dissipation therefore presents a real limitation on the packing density of integrated devices, and the high packing density of 10⁸ stages per cubic foot cannot be maintained over a volume of, for example, one cubic foot with the devices described unless special precautions are taken. For spherical symmetry, the volume-tosurface ratio goes up linearly with size of the integrated device. The larger the device, the lower the packing density has to be.

While the limitation of physical size and that of temperature may be expected to cause trouble at the higher packing densities of future devices, the next limitation to be investigated is felt already in present devices. This is the limitation caused by the probability of errors in the fabrication process leading to shrinkage, and will be treated in the next section.

III. SHRINKAGE CONSIDERATIONS

A. Influence of Shrinkage Rate

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10-1

10⁻²

10⁻³

104

10

10

As will be shown in this section, the main difficulty with integrated devices is sensitivity to shrinkage in the fabrication process. The reason for this is that, as a rule, each of the units in an integrated device must function properly. Thus the rejection of a few inferior or faulty units out of a group, which is possible when indi-

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vidual units are used, is seldom possible in the case of integrated devices, and the entire unit may have to be discarded. An analysis of this situation shows that, for an integrated device to be competitive, the shrinkage per component should not exceed a limit which depends on the number of components and the way in which they are combined. This requirement in general is rather stringent.

For the purpose of comparison between individual units and integrated devices, shrinkage may be divided into two main categories. The first category is shrinkage caused by uncorrelated accidents. Here the chance that a unit is lost is independent of what happens to other units. Such accidents may be the breaking of a wafer, dropping a unit on the floor, contaminants on a unit, etc. The second category is shrinkage caused by correlated accidents. Here the fate of one unit is the same as, or related to, that of other units. Such accidents may be the dropping of a batch on the floor, faulty raw material affecting many units, operator inability, etc. Shrinkage falling in the second category affects individual units and integrated devices equally in the limit when all units in a batch are rendered useless and will be neglected as a favorable case giving no excess cost of integrated devices. Shrinkage falling in the first category, however, represents the worst possible case with regard to integrated devices, and will therefore be considered in detail. In reality, shrinkage is a compromise between these two extremes.



COMPOSITION

Fig. 5—Ratios of active to total volume for typical electronic components.

TRANSISTOR

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B. Temperature Limitation

Now let us turn to the limitation on packing density imposed by heat dissipation. In a binary computer application all the elements would be either ON or OFF at all times. In the most unfavorable case, all the elements would be ON, in which case the power per element may typically be 0.01 watt. The temperature in a sphere consisting of identical elements with volume d^3 , each developing P watts, is given by

$$\theta = \theta_0 + \frac{P(R^2 - r^2)}{6\lambda d^3}, \qquad (1)$$

where

 θ is the temperature at radius r,

 θ_0 is the temperature at the surface at radius R,

 λ is the heat conductivity.

For integrated devices, consisting of layers of semiconductor between printed circuit ceramic wafers, the heat conductivity would be a compromise between that of air (0.025 w/m C°), that of semiconductor (Si 148 w/m C°, Ge 58 w/m C°), that of ceramic ($\approx 2 \text{ w/m C}^\circ$) and that of metal (Sn 65 w/m C°). A fair estimate of the resulting heat conductivity may be 1 w/m C°. With silicon, $\theta_0 = 50^{\circ}$ C, $\theta_{max} = 175^{\circ}$ C, $d = 5 \times 10^{-4}$ m, *R* is obtained as 3×10^{-3} m corresponding to 6 layers or about 200 elements.

This may also be seen from Fig. 6, which shows the number of elements in a sphere N vs P, with d and R as parameters. Although there are well-known methods to raise this temperature limitation considerably, such as the use of high band-gap semiconductors (gallium arsenide, indium phosphide, etc.), design of the devices with a high surface-to-volume ratio to allow more efficient cooling, etc., it appears advantageous from a heat-conduction point of view to limit the size of integrated





sistor. A voltage applied to the end contacts results in a current through the silicon bar and, consequently, an output signal if all the four gate-regions are simultaneously, and only slightly, reverse-biased. If, on the other hand, one or more of the gate regions are strongly reverse-biased, no current will flow along the bar and no output signal is obtained. The use of unipolar transistors makes possible the construction of a large variety of different computer circuits in integrated form.

B. Characteristic Features of Integrated Devices

The main advantage of integrated devices is a considerable saving in space and weight. The devices described above have a packing density of about 10⁸ components per cubic foot. This very large packing density results not only from the integrated device technique proper, but also from the assembly of a large number of bare, nonencapsulated units and circuits within the same enclosure, from the incorporation of printed circuit techniques even in the device itself, from sandwiching such printed circuits, and from standardizing the active and passive components to identical shape and size.

In this way, it has been possible to approach miniaturization in a functional manner, reminding one of the functional trend in architecture some 30 years ago. Nonfunctional details, such as air-spaces between elements and excess semiconductor, have been reduced to a bare minimum. Structural support, which usually has to be provided to all parts, e.g., the connecting wires, is supplied by the semiconductor and the insulator wafers with simultaneous electrical, and sometimes encapsulating, functions. Internal metallic connections are eliminated, or may be reduced to minimal cross sections dictated only by electrical requirements. An illustration of this tight packing is given in Fig. 3, which contrasts a bistable stage of 1948 with a similar stage of 1958 and a similar stage plus delay line in integrated form. (Some of the reduction in size is obtained by simplified, and therefore more critical, operation in the lastmentioned case. Also, encapsulation is not shown.)

II. LIMITATIONS OF INTEGRATED DEVICES

A. Physical Size Limitation

Now let us turn to some of the limitations of integrated devices. The first is a rather practical one, namely that of physical size or packing density. The packing density of some existing electronic components and some simple systems is shown in Fig. 4. It is assumed that the components are packed as densely as possible, and that leads, tube pins, etc., are disregarded. The smallest single component commercially available at the present time is the miniature resistor, and it is very unlikely that much smaller individual components would be made even if this were technically possible. In the miniature resistors, the leads constitute approximately 50 per cent of the volume, and they are difficult to handle in their present form, e.g., picking them up from a flat surface without the help of tweezers, etc. Clearly, there must be a point of diminishing return, where future reduction in size makes it impractical to handle each individual component without adding some form of handle, for instance in the form of leads. When these handles become the largest part of the component, further reduction in size appears pointless.

One way out of this dilemma of not being able to reduce the size of components for loss of handling ability, is to use the components in groups, arrays or sheets, combining several elements into superstructures which then are of such a size that they can be handled. For a moderate number of elements, this represents the middle area in Fig. 4. The integrated devices described in this article fall in this domain.

Extending the argument still further, there must be a limit at which even arrays of moderate complexity become so small that they again cannot be handled without special tools. Beyond this comes an area, corresponding to the top part of Fig. 4, where it is not possible to handle less than an entire system, unless special tools, as yet completely unknown to electronic manufacturing, are used. It is not surprising to find the human brain in this category, and it may be illuminating to think of the difficulties encountered in experimental studies of the electronic activities of the brain in order to appreciate the breakthroughs needed to make sensible use of a packing density exceeding that of the brain. It appears that, for the foreseeable future, the human brain represents a ceiling for packing density which will not likely be surpassed with any presentlyenvisaged technique.

Another factor to consider in designing microminiature electronic systems is the relative volume efficiency of electronic components, which is illustrated in Fig. 5. The figure shows the ratio of electronically active volume to total volume for some common electronic components. In order to bring this figure close to one, encapsulation of single components must be discarded in favor of encapsulation of circuits or entire systems, mechanical support must be minimized or shared, etc.

Fig. 3—Left, bistable stage as made in 1948; center, bistable stage as made in 1958; right, bistable stage and delay line in integrated design (at pencil point).



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Design Considerations for Integrated Electronic Devices*

I. T. WALLMARK[†]

Summary-Some fundamental factors affecting the design of integrated electronic devices are discussed, particularly the influence of shrinkage. It is concluded that the considerable advantages of integrated devices, compared to conventional devices, such as very small volume and weight, and reduced number of metallic connections, have to be paid for by higher shrinkage in fabrication. Three considerations are advanced, which will reduce partially this higher shrinkage. First, the resulting increase in cost may be made very small if the design of the integrated device allows the extent of integration to be adjusted to the shrinkage rate. Second, the resultant high cost of the integrated device justifies a higher investment in the fabrication process of the integrated devices than for the individual units. Third, methods of doctoring integrated devices may be used to reduce the shrinkage effectively.

1. DESCRIPTION OF INTEGRATED SEMICONDUCTOR DEVICES

.1. Specific Examples, Shift Register and Unipolar Transistor Logic

ET us first consider some practical embodiments of integrated devices. By an integrated device or circuit is meant, following the generally accepted terminology,¹ a device consisting of one piece of solid into which have been integrated several component functions, active as well as passive, without external interconnections.²⁻⁵ Two recently developed constructions will be used as examples, namely an integrated shift register⁴ and the integrated direct-coupled unipolar transistor logic.5

The *integrated shift register* which is shown in Fig. 1 consists of a number of bistable stages, each similar to a thyristor.⁶ The individual stages are interconnected by small sections of semiconductor, each serving as a delay line. Each stage has a load resistance in series, and may be in either of two stable states; ON, corresponding to a high current in the stage, or OFF, corresponding to a low current in the stage. These states may be propa-

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¹ As used by I. M. Ross in a review "Functional Devices," at the Electron Devices Meeting, Washington, D. C.; October 29-31, 1959 ² L. A. D 'Asaro, "A stepping transistor element," 1959 WESCON CONVENTION RECORD, pt. 3, pp. 37-42

J. S. Kilby, "Semiconductor solid circuits," Electronics, vol. 32,

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⁵ J. T. Wallmark and S. M. Marcus, "Integrated devices using direct-coupled unipolar transistor logic," IRE TRANS. ON ELECTRONIC COMPUTERS, vol. EC-8, pp. 98-107; June, 1959.
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gated along the bar by the application of a shift voltage to the two end contacts. In this manner, a binary number, stored in the register, may be shifted through the register. Each stage is also provided with a base contact for write-in of information. The semiconductor bar between the stages stores the minority carriers during the shift process and serves as a short-time memory. This function is usually carried out by RC circuits in conventional registers. In effect, therefore, the circuits connecting the different stages have been integrated into the device. The delay line itself would be characterized, according to the generally accepted terminol ogy_{i}^{1} as a functional device. A functional device uses an electronic phenomenon, in this case minority carrier transfer, to accomplish a circuit function. Thus the components integrated in a functional device cannot be separated.





Fig. 2-Multiple AND gate using direct-coupled unipolar transistor logic.

Another integrated device, namely a multiple AND circuit using direct-coupled unipolar transistor logic is shown in Fig. 2. It consists of five unipolar transistor elements fabricated in one piece of silicon, four of which serve active device functions, while the fifth performs a passive component function, in this case that of a re-

Scanning the Issue.

Design Considerations for Integrated Electronic Devices (Wallmark, p. 293)-Integrated electronic devices will probably be the next major step in the evolution of electronics. By incorporating the functions of several components in one tiny solid, it will be possible to achieve packing densities of about 100 million components per cubic foot, resulting in a drastic saving in space and weight. There are, however, a number of practical limitations which will have to be faced. This paper explores the most important present-day difficulty with integrated devices, namely, the decreased number of usable units that are liable to be produced during the fabrication process. This decrease, called shrinkage, is due to the fact that an integrated device is in essence a multi-unit device. If any one of these units is faulty, the whole device will probably have to be discarded. This shrinkage factor, as the author shows, plays a very important role in determining the maximum number of units it is economically feasible to integrate into a single device. The author goes on to develop several methods that will partially reduce this high shrinkage factor to which integrated devices are prone.

Perceptron Simulation Experiments (Rosenblatt, p. 301) -Since 1957 the Cornell Aeronautical Laboratory has been conducting a series of unusual pattern-recognition experiments with a theoretical model of a rudimentary brain, called the perceptron. The perceptron is a simplified, but biologically plausible, net of neural elements which theoretical studies have shown is capable of learning to discriminate and to recognize perceptual patterns. This capability has now been verified and further explored by using a computer to simulate the behavior of the perceptron. The visual stimuli which are generated by an image falling on the retina of an eve are represented by dot patterns which are fed to the computer. The action of every cell and connection in the perceptron brain model is then carried out by the computer. The output response indicates whether or not the input image has been correctly identified. Using alphabetic characters and various geometric shapes as input images, the author ran several interesting experiments designed to study how rapidly and how well the perceptron could 1) be taught and 2) spontaneously learn to discriminate between and recognize different patterns. The results provide valuable data on learning rates and capabilities under various circumstances and provide insight into conditions under which different systems break down or deviate from typical biological learning phenomena. Above all, the experiments go a long way toward demonstrating the feasibility of a perceptron as a pattern-recognizing device.

Pulse Compression-Key to More Efficient Radar Transmission (Cook, p. 310)-Modern military requirements for radar and the advent of radar astronomy have placed much emphasis on methods of increasing the ranges at which objects may be detected. Radar ranges can be extended by increasing average power by means of a wider transmitted pulse, but only with an accompanying, and often unacceptable, sacrifice in range resolving capability. Consequently, designers have turned to stepping up the peak powers of tubes and, further, employing post-detection integration techniques, leading to an inefficient use of tubes and of available average power. This paper discusses an important proposal for increasing the average power capability of a pulse radar without increasing the peak power output of tubes or degrading the pulse resolution. The technique consists of linearly sweeping the carrier frequency of the pulse to be transmitted and of feeding it through a time-delay filter whose frequency characteristics are such as to delay one end of the pulse relative to the other. The result is to compress the width of the pulse and to increase its amplitude, but without exceeding the actual peak power limitations of the system.

Field Effect on Silicon Transistors (Schwartz and Levy,

p. 317)—By observing what happens to the alpha of a silicon transistor when dc electrostatic fields of various strength are applied to the surface, the authors have developed a direct method of calculating surface recombination velocity. This work ties surface recombination theory directly to an operating device, making an excellent addition to the literature on field-effect surface studies.

A Half-Watt CW Traveling-Wave Amplifier for the 5-6 Millimeter Band (McDowell, et al., p. 321)—The production of a half watt output over a bandwidth of 10,000 mc centered at 55,000 mc represents a major advance in the state of the millimeter wave art, exceeding previously obtainable output powers in this frequency range by at least ten to one. As the first practical CW power amplifier with broadband performance in the millimeter range, this development has broad implications for the future, especially in connection with the eventual use of circular waveguide systems for very broadband millimeter communications. In addition, the techniques of helix mounting and cooling will be of great interest to tube engineers,

The Propagation of Radio Waves of Frequency Less Than 1 KC (Pierce, p. 329)—This paper is concerned with propagation phenomena that occur at frequencies as low as 100 cpsprobably the lowest radio wave frequency that has ever been discussed in the PROCEEDINGS. This represents the low end of a frequency band that is now the center of much attention. The authors have found discrepancies between the observed attenuation of signals in the 100 cps to 1000 cps range arising from lightning discharges and the attenuation predicted by theory, which visualizes propagation at extremely low frequencies as occurring in the fashion of a mode traveling within a waveguide formed by the earth and a concentric ionosphere of constant height. It is shown that the discrepancies may be explained by replacing the constant-height model with an ionosphere which increases in height as frequency decreases. Perhaps equally important, this work shows that further experimental observations are needed at these extremely low frequencies.

Maximum Stable Collector Voltage for Junction Transistors (Schmeltzer, p. 332)—When a transistor is operated at high voltages, its stability may be jeopardized either by current multiplication or thermal heating effects, leading to premature breakdown. The author makes an excellent analysis which takes both effects into account quantitatively, thereby providing a basis for specifying a biasing voltage that is safe. In an example he shows that although the breakdown voltage due to multiplication effects alone is 54 volts, and due to thermal effects alone is 53 volts, the interaction of the two lowers the maximum stable voltage to 37 volts.

Ionospheric Models as an Aid for the Calculation of Ionospheric Propagation Quantities (de Voogt, p. 341)—A large computational program of ionospheric ray tracing on a computer is summarized in this brief, readable paper. The results are given in the form of curves which are very useful to people involved in ionospheric studies and in predicting communication frequencies. Although model calculations of this sort are not new, this paper is particularly interesting because it uses fresh models much more in line with what we now know about the distribution of electron density in the ionosphere from rocket investigation.

Electromagnetic Properties of High-Temperature Air (Bachynski, *et al.*, p. 347)—This paper gives an interesting summary of the electromagnetic properties of air at high temperatures and, as an illustrative application of the theory, an estimate is made of the attenuation and phase delay of radio waves from a hypersonic vehicle at various speeds and altitudes in space.

Scanning the Transactions appears on page 411.

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Harry Nyquist Winner of the IRE Medal of Honor

Harry Nyquist (A'39-M'47-F'52) was born on December 7, 1889 in Nilsby, Sweden. He attended the University of North Dakota, Grand Forks, from 1912 to 1915 and received the B.S. and M.S. degrees in electrical engineering in 1914 and 1915, respectively. He attended Yale University, New Haven, Conn., from 1915 to 1917, and was awarded the Ph.D. degree in 1917.

From 1917 to 1934 he was employed by the American Telephone and Telegraph Company in the Department of Development and Research Transmission, where he was concerned with studies on telegraph picture and voice transmission. From 1934 to 1954 he was with the Bell Telephone Laboratories, Inc., where he continued in the work of communications engineering, especially in transmission engineering and systems engineering. At the time of his retirement

from Bell Telephone Laboratories in 1954, Dr. Nyquist was Assistant Director of Systems Studies.

During his 37 years of service with the Bell System, he received 138 U. S. patents and published twelve technical articles. His many important contributions to the radio art include the first quantitative explanation of thermal noise, signal transmission studies which laid the foundation for modern information theory and data transmission, the invention of the vestigial sideband transmission system now widely-used in television broadcasting, and the well-known Nyquist diagram for determining the stability of feedback systems.

Since his retirement, Dr. Nyquist has been employed as a part time consultant engineer on communication matters by the Department of Defense, Stavid Engineering Inc., and the W. L. Maxson Corporation.



Haraden Pratt

Winner of the IRE Founders Award

Haraden Pratt (A'14-M'17-F'29), Secretary and Past President of the IRE, was born in San Francisco, Calif., on July 18, 1891. He began his radio career as an amateur in 1905, and from 1910 to 1914 was a wireless telegraph operator and installer of equipment for the United Wireless Telegraph Company and Marconi Wireless Telegraph Company of America.

In 1914 he received the B.S. degree in electrical engineering from the University of California, and thereafter became a construction and operating engineer for the Marconi Company's trans-Pacific radio stations in California.

As an Expert Radio Aide for the Navy Department from 1915 to 1920, he was concerned with the construction and maintenance of its high-powered radio stations. In 1920, he began the establishment of the public service radiotelegraph system of the Federal Telegraph Company on the West Coast. In 1925 he constructed and operated a radiotelegraph system between Salt Lake City and Los Angeles for the Western Air Express, the first air mail contractor in the western United States. Later he was in charge of development work on radio aids for air navigation at the National Bureau of Standards. In 1928 he became Chief Engineer, and later Vice-President, of Mackay Radio and Telegraph Company. He constructed its world-wide communication system.

For his work during World War 11 as Chief of the National Defense Research Committee's Division 13 on Communications. Mr. Pratt was awarded a Presidential Certificate of Merit. Immediately after the war, he became Vice-President and Chief Engineer of the Commercial Cable Company, All America Cables and Radio, Inc., and the American Cable and Radio Corporation. For many years he held offices in other companies of the International Telephone and Telegraph Corporation, but retired from these activities in 1951. In October of that year, he received a Presidential appointment to the newly-created post of Telecommunications Advisor to the President. He has since retired from government service, and more recently has engaged in consulting services. For a period of twenty-four years he was a member of the United States delegations to international radio and telecommunications conferences.

Mr. Pratt is a member of Sigma Xi. As life member of Veteran Wireless Operators Association, he was awarded the Marconi Medal of Achievement in 1951. He is a Fellow of the American Institute of Electrical Engineers and the Radio Club of America, an Associate Fellow of the Institute of the Aeronautical Sciences, and an honorary life member of the IRE, Australia. In 1944, he received the IRE's Medal of Honor. **Proceedings of the IRE**



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Poles and Zeros



Ides Plus Six. The Romans had no name for this date (March twenty-first) but the members of the IRE do. The date marks

the beginning of another National Convention and Radio Engineering Show. If the Romans had been fortunate enough to have had such an occasion they would have referred to it as 'dies faustus,' the French would describe it as an 'occasion magnifique,' and Hollywood as a stupendous, colossal, and mammoth extravaganza. In plain English—60,000 serious, earnest, professional people meeting to discuss, to consider, and to see the latest advances in their chosen art is an event of major significance and does not require descriptive hyperboles!

What a program of papers and exhibits! Two hundred and fifty-eight papers will be presented; a rough approximation of the information content based on this number of papers, the length of the presentations, and an estimate of the average reading rate, yields 10 million "bits." Eighty per cent of the production capacity of the electronics industry will be represented by 850 exhibitors, with 1000 plus exhibits, and more than 20,000 items. Fifty-four sessions have been organized with the assistance of the Institute's twenty-eight Professional Groups. This is a vivid demonstration of the importance of the Professional Group structure to the dynamics of the Institute. A careful perusal of the program in this issue is recommended. You will find sessions and features of specific interest.

A Panel Session "Electronics—Out of This World" will be a unique, special feature of the program. The Seminar on the 1959 ITU Geneva Conference will present a panel of high officials from the FCC, the State Department, and ITU. Lloyd V. Berkner, President of Associated Universities, Inc., will address the Annual Meeting.

International once more. From time to time, Poles and Zeros has emphasized (and will undoubtedly continue to emphasize) the international aspects of the IRE. There are 6000 members and 22 sections in 84 countries outside the United States. Convention attendance in 1959 from outside the United States reached 1000, representing 38 different countries. In recognition of the truly international character of the Institute, the Board of Directors in meeting assembled, on January 6, 1959, amended By-Laws Section 701.1 so that hereafter our annual gathering will be known officially as the "IRE International Convention and Radio Engineering Show." We welcome the new name as a further step in just recognition of the international scope of the Institute.

Keeping abreast. Our passion for the written word is seldom stated. It behooves all of us to know of the availability of information pertinent to our particular area of interest. In the last several years we have become increasingly aware of the

necessity of familiarizing ourselves with what the Russians are publishing. There has been a profusion of translation services attempting to meet this particular need. As an accommodation to the members, therefore, the PROCEEDINGS with this issue is inaugurating a new listing. The sources of the available English translations of the Russian literature, in so far as we are aware of them, will be printed at the end of the Abstracts and References section each month.

While investigating these Russian translation services, the Editor also discovered sources (unknown to him) of information on publications calculated to assist in keeping up-to-date with United States authors as well. This month, therefore, we include, also at the end of the Abstracts and References section, the story of the facilities of the Office of Technical Services of the U.S. Department of Commerce. Each month, OTS provides over 100 new reports of electronics research and development. The reports range through development and application of new or improved transistors, diodes, tubes, amplifiers, circuitry, and antennas, to research on entirely new electronic systems. The reports are the product of research by the Army, Navy, Air Force, Atomic Energy Commission, and other agencies who turn over their unclassified information to OTS for reproduction and distribution. OTS also distributes translations of Russian and other technical literature which it collects from agencies of the Federal Government. Having spoken, we take no responsibility for your failure to read it all!

Know your Institute. On the contents page of each issue of the PROCEEDINGS there will be found the complete list of the officers and directors of the Institute. It seems appropriate to call this to your attention now so that members may become acquainted with the 1960 roster; it appeared for the first time last month and will, of course, continue throughout the year. To further inform the membership on the development of the Institute, appropriate statistics will be published each month. These statistics will summarize the latest membership and other organizational data. This month these statistics will be found on page 14A.

Congratulations to AIP. Distributed as a supplement to the December 1959 issue of "Physics Today," the second edition (1959) of the American Institute of Physics "Style Manual" is an outstanding publication. Scientists and engineers have long been criticized (unjustly, of course) for their ineptness in the literary field. From now on physicists, at least those who study this excellent document and heed its guide lines, will have no excuse for being inept. The manual deals with the preparation of a scientific paper; general style, presentation of mathematical expressions, preparation of illustrations, and other pertinent subjects. It is clearly organized and profusely illustrated. Congratulations to the AIP, and in particular to its Publication Board.—F. H., Jr.





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(Continued from page 277A)

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Completely integrated Engineering, Design, and Manufacturing Services to Industry and Govern-ment, Microwave antennas, slip ring assemblies, cavity type tuners, fabricated and cast micro-wave components and assemblies, Mixers, Du-plexers, Bends, Tees, Couplers, Feed Horns, Cavities, Dunmy Loads and Electro-mechanical assemblies. assemblies.

(Continued on page 282A)

Your firm not listed? If you would like to become an exhibitor in the 1961 show, write for information to IRE Exhibits Dept., 72 W. 45th St., New York 36.



Floor Plan—Fourth Floor Production





and north sides of the main at the southeast corner, and the and north sides east Elevators at the east to the "4000 Court" lobby take you direct to this floor. Be sure not to miss the booths in the the "4500 Court." in the northwest corner. floor. fourth the G are located Coliseum All lecture halls in the

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MENTION---PROCEEDINGS

ADVERTISERS PLEASE

WHEN WRITING TO

Whom and What to See at the Radio **Engineering Show**

(Continued from page 275A)

Gertsch Products, Inc. 3211 S. La Cienega Blvd. Los Angeles 16, Calif. Booths 3701-3703 ▲ E. P. Gertsch, ▲ E. W. Watts, R. M. Bioni-arz, H. F. Richardson, H. P. Faris



Model CRB

AC Voltage Standard; Automatic Complex Ratio Bridge; 0001% Accurate VIIF Frequency Meter; Microwave Frequency Multiplier; FM Deviation Meter; Frequency Converter; AC & DC Ratio Standards; Coaxial Switch, Per Cent Deviation, Programmable and Shaft Driven Ratio Transformers, Special Transformers.

Giannini Controls Corp., Booths 1428-1430

918 E. Green St. Pasadena 1, Calif.

Carl E. Calohan

Air Data Instruments dealing with pressure, altitude, true air speed, pressure ratio. Mach Number, temperature probes and vanes, and featuring the Mach Switch; Inertial Instru-ments—accelerometers, gyros; Precision Servo Instruments—potentioneters, pressure switches, stepping motors; Systems-featuring TAS. Thrustmeter, and Precision Voltage Monitor Systems Systems

Glasseal Products Co., Inc., Booth 1517 725 Commerce Rd. Linden, N.J.

Al-xander Anderson, K. G. Crocker, Gus Eichelhardt, Jack Goss, Frank Emmet, Russ Diethert, R. Sidnell, Jack Logan, Fred Peter-son, David Humes

*Featuring the new Glass-to-Copper Transistor Terminations, Complete line of Standard and Terminations, Complete line of Standard and Custom-Designed Vacuum-Tight High Compres-sion and Kovar Glass-to-Metal Scals, Transistor terminations, Diodes, Connectors, Leaders, End-scals, Feed-thrus, Stand-offs, Plus "Scaling only" service

Glass-Tite Industries, Inc. 725 Branch Ave. Providence 4, R.I. Booth 1109

John A. Dodenhoff, Ralph R. Papitto, Frank W. Brakenwagen, Jr., Philip Schumacher, Christian D. Berger, Maurice Grosso, Bernard J. Scorza, Le-roy W. Beier, Milton Paisner, Gramer Yarbrough, Gerard V. Dube

Complete line of glass-to-metal hermetic seals for the relay industries, semicon-ductor manufacturers, terminal-feed-thrus, end seals for the capacitor and transformer manufacturers and crystal bases. "D" series line of hermetically-sealed and plastic connectors.

(Continued on page 279A)

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At ACE, we fully inventory all parts for our complete standard line! And when a pot has to be made from scratch - we cut time there, too. All raw materials are warehoused, and a complete machine shop, including Swiss screw machines, is maintained. Our special prototyping department lops the time off special requirements.

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March, 1960

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 273.4)

General Telephone and Electronics Corp., Booths 1906-1908, 2322-2332, 2415-2425

See: Automatic Electric Sales Corp. & Sylvania Electric Products Inc.

General Time Corp. 109 Lafayette St. New York 17, N.Y. Booth 1726A R. W. Behringer, ▲ M. Lace, ▲ W. P. Byrnes, C. B. Higgins, ▲ Dr. W. C. Anderson



INCREMAG-9 Decade Counter

1NCREMAG—Magnetic connter-divider, use-ful in frequency division, timing, programming, control, logic, memory use, computers, etc. Unit above accepts random frequencies 0-20,000 pps and is nine decade counter. All ontputs available. Industrial or military specifications. Furnished as components or systems.

General Transistor Corp., Booths 1212-1214

91-27 138th Pl. Jamaica 35, N.Y.

Jamaica 35, N.Y. ▲ James Evans, ▲ Howard Peaceman, ▲ Frank Sopchick, ▲ Robert Johnson, ▲ Charles As-kanas, ▲ Ted Liebfried, ▲ Edwin Berlin, Joseph Wright, Stephen Tolles, ▲ James Egan Germanium and Silicon Transistors, Semiconduc-tor Diodes, Bobbinless Precision Wire Wound Resistors, Magnetic Recording Heads (Magne-Head Div.)

Genisco, Incorporated, Booth 3244 2233 Federal Ave. Los Angeles 64, Calif.

Jack Kimble, William Tikanen, Alex Weiss, Milton Slawinski

Milton Slawinski Special motors, electronic networks, a rate-of-turn table, miniature indicating lights, and accelerom-eters. Operating displays include Genisco servo-operated rate-of-turn table with Micro-Rate Acces-sory which permits rates as low as 0.0001° per second, a special 40-pole, wond-rotor synchronous motor, and a working accelerometer display.

Georator Corp., Booths 1230-1232 See: Trayco Associates.

Geotechnical Corporation, Booth 3240 3401 Shiloh Rd.

Garland, Texas Frank E. Gaillard, Jack H. Hamilton, James R. Womack, Robert F. McMurray

Womack, Robert F. McMutray Transistorized Timing System, modular: fre-quency standard, recording time-mark program-mer, precise 60-cps power source. WWV com-parator-stability 1 part in 10⁵ per week. FM Telemetering Subcarrier Discriminator, Ground-Station Voltage Controlled Oscillator. Infrasonic Amplifier-0.25 microvolt noise level, Automatic Chart Reader (Curve-Follower).

(Continued on page 277A)

First and Second floors-Components Third floor-Instruments and Complete Equipment Fourth floor—Materials, Services, Machinery

PROCEEDINGS OF THE IRE March, 1960

For **Printed** Circuitry



UCL PRESENTS **STYLE 1005** MIDGET DC RELAY

Cut your material and direct labor costs with

this small, inexpensive, mass-produced relay for use in printed circuitry where the relay is selfsupporting. Designed for simple plug-in installation.

The Style 1005 Relay is a single-pole, doublethrow relay, light in weight yet capable of withstanding severe operating conditions and rough handling.

TYPICAL APPLICATIONS

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GENERAL CHARACTERISTICS

STANDARD OPERATING VOLTAGES MAXIMUM COIL RESISTANCE 13,000 ohms SENSITIVITY 0.05 watt at standard contact rating; 0.3 watt at maximum contact rating NATION SPDT CONTACT COMBINATION SPDT CONTACT RATING Standard 1 amp.; optional ratings, with special construction, to 3 amps. Ratings apply to resistive loads to 26.5 VDC or 115 VAC MECHANICAL LIFE EXPECTANCY 10,000,000 operations minimum

DIELECTRIC STRENGTH 500 VRMS minimum



Also available with solder lugs in open or hermetically sealed styles. **STYLE 1001**



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Floor Plan—Second Floor Components



WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Be sure to visit the South Room which is on the same level, 70 feet off the main floor in the center of the south wall.

WHEN

Whom and What to See at the Radio Engineering Show

(Continued from page 268A)

General Precision, Inc., Formerly General Precision Equipment Corp., Booths 1501-1511

See: GPL Division, Kearfott Division, Librascope Division, Link Division.

General Precision Laboratory Inc., Booth 1505 See: GPL Division, General Precision, Inc.

See: GPL Division, General Precision, Inc

General Radio Co. 22 Baker Ave. West Concord, Mass. Booths 3201-3208

Booths 3201-3208 A.A. E. Thiessen, A Myron T. Smith, A W. R. Saylor, A.S. W. DeBlois, A P. J. Macalka, A.R. K. Peterson, A.R. A. Boole, A W. R. Thurston, A R. B. Richmond, A C. J. Lahanas, A.R. A. Jokinen, A C. E. Worthen, A M. A. Gilman, A L. J. Chamberlain, A E. F. Sutherland, A K. Adams, A C. W. Harrison, A F. J. Finnegan, R. E. Wilson, A I. G. Easton, A R. W. Frank, A E. Karplus, A G. G. Ross, A P. Bishop, A.D. W. Brown, A J. E. Snook, A J. C. Held, A H. H. Dawes, R. H. Reinstra, A D. B. Sinclair, A A. P. G. Peterson, A M. C. Holtje



*Type 1300-A Beat-Frequency Video Generator

Complete impedance standards and measurements laboratory, with impedance measuring equipment from dc to 5000 Mc; *pulse generators; *videofrequency signal generator for sine-wave, squarewave, and sweep testing; random-noise generator; graphic level recorder; audio and acoustical instruments; automatic voltage regulator; display of design features.

General Resistance, Inc. 430 Southern Blvd. New York 55, N.Y. Booth 3026 ▲ Charles Jasik, ▲ Lawrence Merson, ▲ Rubin Blumkin, Mike Lombardozzi, Sam Freed



Precision Wheatstone Bridge

Precision wheatstone bridge—0.0035 accuracy. Precision wire wound resistors, precision resistance networks, voltage dividers, resistance standards, summing networks (DC and 400 cps), cold junction compensators, special temperature coefficient resistors, analog-digital networks.

(Continued on page 275A)

▲ Indicates IRE member. * Indicates new product.

PROCEEDINGS OF THE IRE March, 1960

Dr. Lucius Cuppington introduces . . . VERNITEL, heart of HOOVER'S new FM/FM

HOOVER'S new FM/FM telemetering system that prolongs the life of FM/FM systems now in use, improving their accuracy by a whole order of magnitude:



Count Vladimir Butts Binswinger shows ...



HOOVER'S new Mixer Amplifier, the palm-sized part of the Vernitel system that helps FM/FM telemetering systems live beyond their income, by prolonging their lives amazingly:



Personalities

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HOOVER ELECTRONICS COMPANY

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Sir Joshua Wormwood Scrubbs offers . . .

HOOVER'S new Millivolt Transistorized Oscillator that eliminates DC amplification from telemetering, allowing fewer and smaller packages and an end to one source of error:





Dr. Herpes Zoster introduces . . .

HOOVER'S new Transistorized Subcarrier Oscillator, for FM/FM telemetering circuits, offering a linearity within 0.3% of band-width and a frequency stability within 1.5%.



See them at the IRE Show, March 21-24... or ask for the literature and specification sheets.



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March, 1960

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PROCEEDINGS

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Rated at from 0.325 to 250 amps, in complete variety of case designs and terminals

Proved performance, low cost, prompt shipment from stock

Sarkes Tarzian's "Designers' Line" silicon rectifiers offer the small size, high efficiency, mounting versatility, and wide range of ratings that can help solve many of your power conversion circuitry problems. Tarzian's realistic prices make these high quality components practical for almost all commercial and military applications.

The 84 types of Tarzian "Designers' Line" rectifiers feature extremely low junction current density to provide maximum reliability and operating life. Their -55° C to $+125^{\circ}$ C temperature range makes Tarzian silicon rectifiers ideal for circuits where ambient temperatures are high and small size is desired. Ratings range from 0.325 to 250 amperes.

Tarzian types are available for immediate delivery in production quantities from factory or warehouse stocks. Complete power conversion engineering service on your rectifier requirements is available at no charge or obligation.

For further information contact your nearest Tarzian sales representative or write to Section 4394D, Semiconductor Division, Sarkes Tarzian, Inc., Bloomington, Indiana.

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SEMICONDUCTOR DIVISION — BLOOMINGTON, INDIANA In Canada: 700 Weston Rd., Toronto 9, Ontario Export: Ad Auriema, Inc., New York City

		peak		max.	amps.	NEG	ATIVE	POS	ITIVE			peak		max.	amps.	NEG	ATIVE	POS	ITIVE
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DC (100° C)		volt- age	RMS	rent	surge 4MS	Tarzian Type	Jedec No.	Tarzian Type	Jedec No.	DC (100° C)		volt- age	RMS volts	rent peak	surge 4MS	Tarzian Type	Jedec No.	Tarzian Type	Jedec No.
		50	35	210	350	5S3N	-	5S3P	-			50	35	900	1500	5W3N	-	5W3P	-
35		100	70	210	350	10S3N	-	10S3P	-			100	70	900	1500	10W3N	-	10W3P	-
	-)	200	140	210	350	2053N	-	20S3P	-	150		200	140	900	1500	20W3N	-	20W3P	-
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		400	280	210	350	40S3N	-	40S3P	-			400	280	900	1500	40W3N	-	40W3P	-
	 	50	35	600	1000	5VAN	1N1165	5VAP	1N1179			50	35	1200	2000	5XAN	1N1263A	5XAP	1N1267A
		100	70	600	1000	10VAN	1N1166	10VAP	1N1180			100	70	1 200	2000	10XAN	1N1264A	10XAP	1N1268A
100		200	140	600	1000	20VAN	1N1167	20VAP	1N1181			200	140	1 200	2000	20XAN	1N1265A	20XAP	1N1269A
		300	210	600	1000	30VAN	1N1168	30VAP	1N1182			300	210	1200	2000	30XAN	1N1266A	30XAP	1N1270A
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		300	210	600	1000	30V3N	-	30V3P	-			300	210	1200	2000	30X3N	-	30X3P	-
		400	280	600	1000	40V3N	-	40V3P	-			400	280	1200	2000	40X3N	-	40X3P	-
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		100	70	900	1500	10WAN	1N1264	10WAP	1N1268	250		100	70	1500	2500	10Y3N	-	10Y3P	-
		200	140	900	1500	20WAN	1N1265	20WAP	1N1269			200	140	1500	2500	20Y3N	-	20Y3P	-
		300 210 900 1500 30WAN 1N1266 30WAP 1N1270			300	210	1500	2500	30 Y 3 N	-	30 Y 3 P	-							
		400	280	900	1500	40WAN	-	40WAP	-			400	280	1500	2500	40Y3N	-	40Y3P	-



efficiency types

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0C (100° C)		inverse voltage	R M S volts	recurrent peak	surge 4MS	Tarzian Type	Jedec No.	amps. DC		in- verse volt-	max. RMS	re- cur- rent	SULTE	Tarzian	Jedec	Tatzian	Jedec
0.325		2800	1960	3.25	19.5	280 S M	1N1113	(100° C)	-	age	volts	peak	4MS	Type	No.	Туре	No.
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0.375	Cr. market	2000	1400	3.75	22.5	200 S M	181111	1,5		200	140	10	100	-	-	20J1	1N1618
0.4		1600	1120	4	24	160SM	1N1110			300	210	10	100	-	- 1	30J1	1 N1619
0.425		1200	840	4.25	25.5	120SM	1N1109	-		400	280	10	100	-		40J1	1N1620
0.45		800	560	4.5	27	80 S M	1N1108	2		100	70	30	100	-	-	10LA	1N1085
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0.5	500	300	210	5	30	30M	1N1083		-	100	70	50	150	-	-	10J2	1N1621
		400	280	5	30	40M	1N1084 10	10		200	140	50	150	_	-	20J2	1N1622
		400	280	5	30	M-500	1N1084			300	210	50	150	-	-	30J2	1N1623
		500	350	5	30	50M	-			400	280	50	150	-	_	40J2	1N1624
		600	420	5	30	60 M	-			50	35	120	200	5RAN	1N1157	5RAP	1N1171
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		400	280	7.5	75	F-4	1N2483			200	140	120	200	20RAN	1N1159	20RAP	1N1173
		600	420	7.5	75	F-6	1N2484			300	210	120	200	30RAN	1N1160	30RAP	1N1174
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0,75	- (B)	200	140	7.5	75	20H	1N2485	N 2485		50	35	210	350	5SAN	1N1161	5SAP	1N1175
		300	210	7.5	75	30H	1N2486		100	70	210	350	IOSAN	1N1162	10SAP	1N1176	
		400	280	7.5	75	40H	1N2487	35		200	140	210	350	205AN	1N1163	20SAP	1N1177
		500	350	7.5	75	50H	1N2488			300	210	210	350	30SAN	1N1164	30SAP	1N1178
		600	420	7.5	75	60H	1N2489			400	280	210	350	40SAN	-	40SAP	-





NEW FROM JFD LUMPED CONSTANT DELAY LINES

Meet the newest addition to the growing family of JFD precision electronic components.

Designed with compactness, ruggedness and reliability in mind, new JFD lumped constant Delay Lines upgrade your prototype or production project.

Compare the advantages of the standard JFD lumped constant delay lines:

- High delay-to-rise time ratio with minimum signal attenuation.
- Tolerance of $\pm 5\,\%$ max, on delay and characteristic impedance.
- Temperature range of -55° C to $+125^{\circ}$ C.
- Delay time thermal stability of 50 parts per million per degree centigrade.
- Up to 25 Me bandwidth.
- Virtually linear phase shift.
- Hermetically sealed metal cases for maximum resistance to shock, vibration and humidity.
- Meet all applicable MIL specs.

Whether your application calls for standard or custom-built lumped constant or distributed constant delay lines, our engineering staff will be glad to review your needs and

Typical Standard Delay Line Characteristics

Delay	Time 5 # sec.		10 # sec.	25 # sec.				
Time	Size	Rise Time	Size	Rise Time	Size			
1.0	11/8×11 8×21/4	2.0	11/2×11/2×3	5.0	1%6×1%6×2%8			
.5	15/14×15/14×258	1.0	158×158×31/4	2.5	13/4×13/4×31/2			
.3	13/8×13/8×234	.6	13/4×13/4×31/2	1.5	2716×2716×47/8			
15	21/4×21/4×41/2	3	21/4×21 4×41/2	.75	23/4×23/4×51/2			

Range of characteristic impedance: 50 ohms to 2000 ohms $\pm 5\%$.

 \pm 5%. Attenuation: Less than 1db per, # sec, up to 3 # sec, delay; 6db max, up to 50 # sec, delay. Temperature stability: 50 parts per million per degree C from -55° to +125° C.

submit recommendations. Closer tolerance delays and impedances are available, in forms. sizes and terminal designs to match your needs. Write for Bulletin No. 213A.



Microwave Cavities

- Oscillators
- Filters
- Frequency Meters
- Frequency Standards

Whether you require the use of S band or C band triode transmitter and receiver oscillators, or tunable filters, John Gombos Co., Inc. provides complete design, manufacture and test of these and all other types of microwave assemblies, including sub-systems.







See us at the I.R.E. Show Booths 2342-2344

Whom and What to See at the Radio Engineering Show

(Continued from page 266A)

General Findings and Supply Co. Industrial Div. Attleboro, Mass. Booth 4052

Gerald F. Tucci, Fred Dole, Sam Greenbaum, Arthur O. Marcello, Jr., Peter Microulis, B. Hocker



Miniature precious metal contacts and assemblies. Rivets, Brushes, Slip Rings, Special Shapes, All standard contact Gold, Silver, Platinum and Palladium. Base metal precision fabrications.

General Instrument Corp., Defense & Engineering Products Group, Booths 1218-1224 81 N. 4th St.

Brooklyn 11, N.Y.

▲ Marvin Hobbs, ▲ Philip Lepofsky, ▲ Robert Cartwright, ▲ Eugene Weisberger, ▲ Charles Hittner, ▲ Samuel Sablove, Nathan Borgman, ▲ Edwin Pores, ▲ Murray Shainis System design & production in radar, identification, communication, microwave telephone & telegraph systems, navigation, guidance, beaconry, telemetry, air traffic control, sonar & ultrasonics, airport instrumentation, meteorology, thermoelectric devices, and transducers.

General Instrument Corp., Radio Receptor Co., Inc., Subsid., & Micamold Electronics Mfg. Corp., Div., Booths 1218-1224

See: Radio Receptor Co., Inc. & Micamold Electronics Mfg. Corp.

▲ Indicates IRE member, * Indicates new product.

Lecture Halls in the Coliseum are located on the Fourth Floor. See complete program of speakers and papers in the editorial section of this issue. General Instrument Corp. Semiconductor Div. 65 Gouverneur St. Newark 4, N.J. Booth 1218-1224

BOOM 1218-1224 D. Adler, M. Barmat, H. Chapman, G. Fox, M. Friedman, A. Gartner, V. Griski, S. Gross, S. Gurion, H. Hagler, A. Kosowkoski, M. Lissner, J. Loebenstein, A. Nash, H. Nash, P. Pritchard, C. Schuler, A. Sikorsky, S. Soloman, J. Tucker, S. Winuk

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General Magnetic Corp., Booth 1904 4937 Fullerton Ave. Chicago 39, Ill.

W. E. Gilman, ▲ R. T. Thompson. ▲ P. Gehhardt, R. Campbell, F. Da Roza Magnets, magnetic metals and materials.

> General Mills, Inc. Mechanical Division 1620 Central Ave. N.E. Minneapolis 13, Minn. Booths 3937-3939

▲ Richard Quinn, ▲ Francis Alterman, ▲ Stuart Ulfers, Leonard Kriedemacher, Ted Conant, Lloyd Pearson



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General Motors Corp., Booth 1226 See: Delco Radio Division.

(Continued on page 273.4)

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Whom and What to See at the Radio Engineering Show

(Continued from page 265A)

General Electric Co. Semiconductor Products Dept. Electronics Park Syracuse, N.Y. Booth 2904

Booth 2904 ▲ H. Potter, H. Sweeney, ▲ W. Hall, ▲ N. Sampson, H. Lowry, B. Alexander, H. Taylor, W. Overstreet, H. Hodsdon, P. Hahn, A. Barko, J. Walton, W. Lupton, A. Woolaver, P. Burks, R. Bond, ▲ G. Curtiss, E. Hookway, C. Huyette, W. Halley, R. Olsen, A. Larmann, W Robusto, G. Galliher, D. Morse, T. Burns, T. Loucas, L. Bassett, J. Teahan, R. Rogers, L. Mooney, ▲ H. W. Gebhardt, C. Goodman, D. Hickie



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General Electric Company, Silicone Products Department, Booth 2932 Waterford, N.Y.

J. W. Hawkins, J. S. Hurley, R. Treat, A. E. Horning, P. A. Goodwin, W. J. Dugan, R. A. Winter, G. A. Darsie

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General Electric Company, Specialty Control Dept., Booth 2928 Waynesboro, Va.

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General Electric Company, Voltage Regulator Section, Booth 2924 Pittsfield, Mass.

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(Continued on page 268A)

Show Hours 10 a.m. to 9 p.m. daily Monday through Thursday March 21-24, 1960

World Radio History

WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

Whom and What to See at the Radio Engineering Show

(Continued from page 262A)

General Electric Co., Missile and Space Vehicle Dept., Booth 2920 3198 Chestnut St. Philadelphia 4, Pa. William Hoese Military Electronics

General Electric Co. Power Tube Dept. Schenectady 5, N.Y. Booths 2912-2914

▲ E. C. Numrych, & G. W. Her, E. A. DeMetre, E. T. Chace, \blacktriangle W. F. McKeehan, R. H. Mack, \blacktriangle C. G. Lob, R. R. Rottier, D. Hodges, E. C. White A. C. Rowe, A. Michaelson, \blacktriangle C. Karabats, K. E. Wilson, C. Vignola, H. L. Clark, \blacktriangle Dr. M. Weinstein, \blacktriangle H. Hannam, W. G. Granat, W. J. Pohl, \blacktriangle Dr. P. Wargo



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General Electric Co. Receiving Tube Dept. Owensboro, Ky. Booths 2906-2908

Booths 2906-2908 W. H. Clarke, ▲ L. T. Bowles, ▲ A. F. Dickerson, R. D. Kennedy, W. F. Greenwood, I. D. Daniels, ▲ E. H. Fritschel, ▲ J. R. Sommerville, ▲ A. P. Haase, ▲ R. E. Moe, C. A. Richardson, G. E. Burns, H. B. Nelson, J. E. Nelson, J. W. Gross, C. D. Cillie, R. A. Kittell, K. K. Krehbiel, Gordon Borgora, E. J. La-Croix, Jr., T. B. Jacocks, A. F. Bohner, R. H. Leach, C. L. Barnette, ▲ E. L. Davis, ▲ Lloyd Mumford, ▲ Frank Snyder, ▲ Bruce Angwin, Robert Hughes, George Crossland, Meredith Hamilton, Ken Weitzel



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(Continued on page 266A)

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PROCEEDINGS OF THE IRE March, 1960

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TEST	UNITS	CTP 1728	CTP 1735	CTP 1729	CTP 1730	CTP 1731	CTP 1736	CTP 1732	CTP 1733
Min BVcbo @ 2 ma	volts	40	60	80	100	40	60	80	100
Min BVceo @ 500 ma	volts	25	40	55	65	25	40	55	65
Min BVces @ 300 ma	voits	35	50	65	75	35	50	65	75
Max Icbo @ 85 C @ Max Vcb	ma	7	7	7	7	7	7	7	7
Typ. Icbo @ 2 V	μa	20	20	30	30	20	20	30	30
D. C. Current Gain @ 0.5A		30-75	30-75	30-75	30.75	60-150	60.150	60-150	60-150
Max Veb @ 3.0 A	volts	1.0	1.0	1.0	1.0	1.0	1.0	1.0	1.0
Max Vce (sat) @ 3.0A, 300 ma	voits	0.7	0.7	0.7	0.7	0.7	0.7	0.7	0.7
Min fae @ 1.0 A	kc	15	15	8	8	10	10	6	6
Max Thermal Resistance	c/w	2.5	2.5	2.5	2.5	2.5	2.5	2.5	2.5

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IRE BOOTH 3033

Whom and What to See at the Radio **Engineering Show**

(Continued from page 260A)

General Electric Co. Heavy Military Electronics Dept. Syracuse, N.Y. Booth 2921 ▲ R. J. Brown, ▲ G. D. Prestwich, ▲ T. F. MacCoun, ▲ N. R. Bibko, ▲ J. J. Schoebel, ▲ G. T. Wolfe, D. T Hambleton



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General Electric Company, Industrial Heating Dept., Booth 2932 Shelbyville, Ind.

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A complete listing of all registrants at the IRE National Convention and Radio Engineering Show, showing company affiliation and local hotel, is posted on the back wall of the mezzanine on the west side of the first floor.

General Electric Co. Instrument Dept. West Lynn, Mass. Booth 2928 J. E. McQuillan, E. R. Harrison, W. E. Corn-ish, W. J. Kearney, S. R. Sulis, M. W. Vittum, R. B. White, J. D. Henderson



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General Electric Co., Laminated Products Dept., Booth 2914 Coshocton, Ohio

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General Electric Co., Large Lamp Dept., Booth 2916

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Tungsten and Molybdenum metal products, electronic phosphors, lead wires and lamp bases along with pilot and indicator lamps.

General Electric Company, Magnetic Materials Section, Booth 2920 Erdmore, Mich.

Erdmore, Mich. M. E. Hartl, R. J. Studders, A. Marquis, R. J. Parker, R. P. Smith B. H. Fellows, G. D. Barcus, Jr., J. C. Betts, J. E. Foy, G. F. Mus-sen, J. P. Young, M. L. Chater, G. J. Anderson, G. R. Laughlin, H. Schwartz Complete line of Alnico Permanent Magnets and typical applications. Thyrite B Varistors for voltage surge suppression. Thermistors for tem-perature compensation, measurement, and con-trol. *New Positive Temperature Coefficient Thermistors. Thermistors.

(Continued on page 265A)



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RMC

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RMC

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 258A)

General Ceramics Div. Indiana General Corporation Crows Mill Rd. Keasbey, N.J. Booths 1310-1312 J. P. Manley, C. L. Snyder, J. W. Schallerer, R. E. Warren, N. Shapiro, H. Landsberger



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General Chemical Div., Booth 4216 See: Allied Chemical Corp.

> **General Dynamics Corp.** Liquid Carbonic Division 135 South La Salle St. Chicago 3, Ill.

> > Booth 3060 C. B. Brisendene

Low Temperature Test Equipment.

General Electric Company, Capacitor Department, Booth 2928 Hudson Falls, N.Y.

F. R. Flood, J. P. Holioway, F. L. Johnson, ▲ W. H. Roberts, D. F. Schmidt, J. G. Hanan, W. C. Bakes, J. Feininger, G. F. Wallin

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(Continued on page 262A)

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World Radio History

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 256A)

Gardner-Denver Company, "Wire-Wrap" Division, Booth 4530 Grand Haven, Mich.

Grand Haven, Mich. H. F. Wilson, G. W. Roth, W. Long, E. Julan-der, W. N. Christiansen, W. F. Bigony, E. L. Keating, R. Van Schelven, C. F. Allen, P. Meaden, B. V. Hayes R. E. Shute, G. P. Shel-don, P. Swayze, J. W. Kaar, G. R. Dana Air and electric "Wire-Wrap" tools, mechanical "Wire-Wrap" tools, unwrapping tools for sol-derless, wrapped electrical connections, Also, wrapping bits and sleeves, Air-powered screw-drivers and nutsetters for assembly of small parts. parts.

Garlock Electronic Products, The Gar-lock Packing Co., Booths 2814-2816 Camden, N.J. and Palmyra, N.Y.

▲ John P. Dearie, Richard Graeff, Friderick O. Dutton, Harry S. Stott, ▲ Parker Naudain, ▲ Karl Bohaker, Joseph Stewart, George Haw-kins, Peter Woodams, Bern Kalp, Wayne Un-derwood, Joseph Burgin, T. L. Denney, R. A. Lyons

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R. Dale Moyer, Richard Callison, Karl Gran-lund, Peter Depp, Hilliard Davis, Charles Baugh, Gerald Rennerts

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Gates Radio Co., Booth 3515-3517 123 Hampshire St. Quincy, Ill.

Larry Cervone, Abe Jacobwitz, Bud Ayer, Bob Hallenbeck, Don Udey Radio Broadcast equipment for AM, FM, TV associated communications transmitters and equipment

General Aniline & Film Corp., Booth 4133

See: Ozalid Division.

General Cement Manufacturing Co., Booth 2126 See: G-C Electronics Mfg Co.

(Continued on page 260.1)

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2N1495	TO-9	250mw	-30v	0.35v	0.60v	60	320mc
2N1496	TO-31	*0.5w	-30v	0.35v	0.60v	60	320mc
2N1204	TO-9	250mw	-20v	0.35v	0.60v	60	320mc
2N1494	TO-31	*0.5w	-20v	0.35v	0.60v	60	320mc

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 255A)

G-L Electronics Co., Inc., Booth 1916 2921 Admiral Wilson Blvd. Camden 5, N.J.

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219 S. Mednik Ave.

Los Angeles 22, Calif.

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GPL Division, General Precision, Inc., Booth 1505

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E. Bernstein, J. Lampson, L. Smith, T. Price, A. Roman, E. Manzo, R. Conkwright, J. Ryan, S. Thomas, A. Anderson, N. Marshall, A. Brundage

Airborne Doppler Radar, Self-Contained Navigation Systems and Track Navigation Com-puters, Closed-Circuit Television; Air Traffle Control, Communications and Data Processing Equipment.

Gabriel Electronics Div., The Gabriel Co., Booths 1720-1722

135 Crescent Rd.

Needham Heights 94, Mass.

▲ Samuel W. Stewart, ▲ Allan W. Jayne, Ed-ward S. Prohaska, Joseph L. Buckley, ▲ Jack B. Hamre, Robert F. Stewart, Bruno Pawlow-ski, Louis Lamperti, J. Thomas Curran

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Gamewell Company, Potentiometer Div., Booth 2838

1238 Chestnut St.

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Potentiometers and miniature metal housed items, including trimmer, precision, differential, multiturn Rotary Switch, and other special potentiometers.

(Continued on page 258A)

First and Second floors-Components Third floor—Instruments and Complete Eqiupment

Fourth floor-Materials. Services. Machinery



Freed Transformer Co., Inc. 1718 Weirfield St. Brooklyn 27, N.Y. Booths 2721-2723 G. T. Dalrymple, M. Salzberg, L. Freed, ▲ D. Gurevics, ▲ S. Solzberg, ▲ R. Freed, R. M. O'Dea, J. Solzberg, M. J. Solzberg



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Freeport Engineering Co., Booth 3947 See: Ebauches $S_{1}A_{2}$

Frequency Standards, Div. Harvard Industries, Inc., Booth 3843B P. O. Box 504

Asbury Park, N.J.

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GEMP Mfg. Corp., Booth 4115 See: Great Eastern Metal Products Co.

(Continued on page 256A)





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Whom and What to See at the Radio **Engineering Show**

(Continued from page 252.4)

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▲ Jack N. Popper, Charlie Kossmann, Bob Latin, ▲ Bernie Gunshinan, Dick Ringer, Bert Aaron

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John Fluke Mfg. Co., Inc., Booth 3242 7100-220th S.W.

Mountlake Terrace, Wash. John Fluke, Roy Malm, Robert Hammond, John Zevenbergen, Leighton Rama, Richard May, Donald Hall Power Supplies; D-C and A-C Differential Voltmeters.

Fluorocarbon Products, Inc., Booths 2814-2816

See: Garlock Electronic Products.

Ford Instrument Co., Booths 2432-2438 See: Sperry Rand Corp.

Foto-Video Laboratories, Inc. 36 Commerce Rd. Cedar Grove, N.J. Booth 3013

▲ Albert J. Baracket, ▲ Thomas R. Kennedy, ▲ Herbert P. Michels, ▲ Charles Halle, Wil-liam Battista, ▲ Robert D. Hamilton, Donald R. Foyer, Joseph J. Kaspar, ▲ Hans H. Nord, Laurence D. Nagy



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Whom and What to See at the Radio **Engineering Show**

(Continued from page 250.1)

Fansteel Metallurgical Corp. North Chicago, Illinois Booths 1021-1022

▲ H. Paul Weirich, J. W. Rose, R. Jaeger, G. Iaggi, W. Bullock, R. Field-man, C. Blanchard, G. Cook, H. Douglas

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Faradyne Electronics Corp., Affiliate of Mansol Ceramics, Booth M-7 471 Cortlandt St.

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Federal Electric Corp., Booths 2510-2520, 2615-2625

See: International Telephone & Telegraph Corp.

Federal Pacific Electric Co., Booths 2725-2727 See: Cornell-Dubilier Electric Corn.



Federal Pacific Electric Co., Booth 2338 See: Fifty Avenue L. Inc.

Federal Shock Mount Corp., Booth 1324 See: Korfund Co., Federal Div.



Fenwal Electronics, Inc., Booth 1102 51 Mellen St.

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Ferris Instrument Co., Booth 3801 110 Cornelia St. Boonton, N.J.

▲ H. J. Tyzzer, ▲ T. S. Leoser, ▲ S. R. Montcalm

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Booth 2530 W. J. Crosby, F. C. Sloboda, ▲ J. E. Moynihan, E. Slaney, Leo Lugten, Peter Geldermans



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Fifty Avenue L, Inc., Div. Federal Pa-cific Electric Co., Booth 2338 50 Paris St.

Newark 1, N.J. L. Van Blerkom, A. G. Lane, E. W. Stohr, G. G. Kahant, H. Pacent, L. Pacent, P. Giroux, P. Piersall

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(Continued on page 254A)

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252A

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 248A)

FXR, Inc. 26-12 Borough PI. Woodside 77, N.Y. Booths 3713-3717 ▲ H. Feldmann, ▲ T. N. Anderson, W. D. Marshall, ▲ J. Ebert, C. T. Zavales, ▲ N. Deoul, ▲ M. Magid, ▲ L. Bertan, R. Diamond, ▲ J. R. O'Donnell



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Fairchild Camera & Instrument Corp., Defense Products Div., Booths 2701-2707 **Robbins Lane**

Syosset, L.I., N.Y.

R. Bruce, D. D. Etkin, E. Paczke, M. Stallone, R. A. Draghi, Jr.

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Fairchild Controls Corp. **Components** Division Subsid. Fairchild Camera & Instrument Corp. 225 Park Ave., Hicksville, L.L., N.Y., & 6111 E. Washington Blvd.,

Los Angeles, Calif. Booths 2701-2707

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New York 3, N.Y.

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Fairchild Semiconductor Corp. Subsid. Fairchild Camera & Instrument

Corp. 545 Whisman Rd. Mountain View, Calif. Booths 2701-2707

T. H. Bay, D. A. Beadling, Gene Keyarts, Richard Lewis, Jr., Donald Rogers, Don Farina, Elmer Biegel, Howard Bobb, Walter Andrews, James Paris, Robert Dugan, Richard Day, Thomas Murphy, William O'Hara



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Booths 1315-1317 R. E. Hill, C. F. Lindholm II, D. R. Contant, W. C. Clapp, H. P. Stuart, J. J. Olah, J. E. Jensen, J. M. Ricker, P. W. Kievit, A. W. Lindholm, C. B. Contant, R. C. Ziegler



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(Continued on base 252.1)

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(Continued from page 245A)

Elgin Metalformers Corp. 630 Congdon Ave. Elgin, Ill. Booths 4420-4424

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Empire Devices Products Corp. 37 Prospect St. Amsterdam, N.Y. Booth 3818-3820 ▲ Michael T. Harges, ▲ Joseph Lorch, ▲ Wil-liam S. Lambdin, W. Leo Bain, Jr.



14 KC to. *Coaxial Ter. & Field Intensity Meters, Noise *Noise & Field Intensity Meters, 14 KC to 10 KMC; Power Density Meter; *Coaxial Crystal Mixers; Microwave Attenuators; Ter-minations; Microwave Power Attenuators; Mi-crowave Attenuator Panels; *Microwave Step Attenuators; *Ultra-Sensitive Receiver; Radar Interference Blanker; Variable Frequency Power Supplies; Microwave Impulse Generators; AM Modulation Meter.

▲ Indicates TRE member, Indicates new product.

First and Second floors-Components Third floor—Instruments and Com-

plete Equipment

Fourth floor-Materials, Services, Machinery



Engineered Electronics Co. 506 East 1st St. Santa Ana, Calif. Booth 1219

▲ T. W. Jarmie, ▲ F. J. Temple, ▲ A. B. Williams, ▲ W. O. Hedge, ▲ K. Goodman, ▲ R. J. Aaron, ▲ R. A. Cepuch, E. R. Lapensee



Packaged Circuit Modules

Transistorized plug-in circuit modules for dig-ital equipment, including basic pulse & logic blocks, Transistor switches, indicating & non-indicating counters, Sensitive indicators, Bread-board panels, Also vacuum tube circuit modules, diode modules, encapsulated solid-state circuits, & power supplies.

Epco Products, Inc., Booth 2239 Wallkill, N.Y.

E. Mullin, Charles Cutney, Charles Abolin, Nat Sperry, Dave Unger

Transformers: Audio; Power; Reactor; Tor-roids;-Transistor Types; Slim Types; Ultra-sonic Units, Manufactured to Military or Com-mercial Specifications. Encapsulated & High Temperature Types.

Epsco, Inc., Booths 3904-3906 275 Massachusetts Ave. Cambridge, Mass.

Peter Zitso

Amplifiers, Automatic Control Equipment, Con-verters, General Test Equipment, Graphic Re-corders, Medical Equipment, Indicating Instru-ments, Printed & Packaged Circuits, Recorders, ments, Printed & Pac Recording Accessories.



Magnetic Components and Assemblies, Slim-Tran Transfor Slim-Tran Transient Filters. Slim-Tran Transformers,

(Continued on page 218A)

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

Electronic Mechanics, Inc., Booth 4305 101 Clifton Blvd. Clifton, N.J.

F. M. Grafton, B. Replogle, K. Ivey, D. Replogle, J. Liker, R. Sachleben Mykroy glass-bonded-mica 700 to 100°F insulation custom molded, machined, and stock shapes extruded, machined and custom molded KEL-F and Teflon insulation specialists.

Electronic News, Booths 4419-4421 See: Fairchild Publications, Inc.

Electronic Representatives Association, ERA Room 319

600 South Michigan Ave. Chicago 5, Ill.

Wally Shulan, ▲ Philip Andress, ▲ R. Edward Stemm, ▲ Harry Halinton, William C. Weber, Ir.

Jr. The trade association of Electronic Manufacturers' Representatives with over 600 members in United States, Canada, Mexico. Featuring a "Lines Available" service to aid manufacturers looking for quality representation to Industrial, OEM, Audio, Distributor, Instrument, and Government Accounts.



D. D. Grieg

D. D. One Transistorized miniaturized power packs, transistorized inverter and converter, nonlinear magnetic components, Slimitron space saving transformer, magnetic amplifiers, special magnetic designs.

Electronic Specialty Co., Technicraft Division, Booth 1818 Thomaston-Waterbury Rd. Thomaston, Conn. John Stinson Microwave Components.

Electronic Tube Corp., Booths 3112-3113 1200 East Mermaid Lane Chestnut Hill Philadelphia 18, Pa. ▲ Kenneth C. Meinken, Sr., ▲ Kenneth C. Meinken, Jr., ▲ Richard T. Rude, ▲ Walter C.

Hill OSCILLOSCOPES: Single-beam; dual-channel; four-channels with plug-in amplifiers, plug-in sweep, "Cali-Marker," delay line, rack-mounted oscilloscopes, CAMERAS: Oscilloscope Continuous Recording Cameras, Oscilloscope Polaroid Land Back Cameras, Oscilloscope Polaroid Land Back Cameras, CATHODE RAY TUBES for industrial and Military-multi- and single-gun tubes for radar and instrumentation.

Electronics International Co., Booth 3018

See: ELIN Division.

Electrosnap Switch Division Controls Company of America 4218 W. Lake St. Chicago 24, Ill. Booths 1727A-1727B George Ballee, Dave Stremmel, Harold Ames, Kace Brin

Ames, kace bin Industry's most complete line of lighted push-button switches (Switchlites). Also, complete line indicator lights, toggles, push-buttons and basic snap-action switches including environment-free and hermetically-scaled models. Custom electro-mechanical assembles include holding coil switches, rotaries, interlocks.

(Continued on page 246.4)

▲ Indicates TRE member. * Indicates new product.





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LONGER LIFE



670

bronze driving pawl and stop

flat phosphor bronze return spring

molded nylon bobbin nylon ratchet and cam

GREATER EFFICIENCY

positive non-jamming stop



Continuous Duty Stepper

NEW

unique features of Guardian's Series 670
 Impulse Relay insure trouble free operation well in excess of one million steps. Each momentary impulse (up to 10 steps per second) causes relay to reverse its cam actuated contacts. Contact arrangements up to D.P.D.T. and ratings to 1500 watts non-inductive, or up to 20 amperes locked motor current, motor load control on 115 volts, 60 cycles. Coil voltages to 230 VAC or 110 VDC.
 Applications include on/off control of lights, motors, appliances and speakers, among others.

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Series 2505 6 P.D.T. Hermetically Sealed Relay



No. 24 A.C. Midget Solenoid



1628-C W. WALNUT STREET, CHICAGO 12, ILLINOIS



(Continued from page 242.4)

Electronic Daily, Booths 4404-4406 Sce: Hayden Publishing Co.

Electronic Design, Booths 4404-4406 See: Hayden Publishing Co.

Electronic Industries, Booths 4301-4303 Chestnut & 56th Sts. Philadelphia 39, Pa.

Robert E. McKenna, ▲ Creighton M. Marcott, ▲ John E. Hickey, Jr., Elmer Dalton, Joseph Drucker, ▲ Menard Doswell, Shelby A. Mc-Million, B. Wesley Olson, ▲ Bernard F. Osbahr, ▲ Richard G. Stranix, ▲ Christopher Celent, Donald J. Moran, Gerald B. Pelissier, George Felt

Electronic Industries "Where the Engineer Comes First," The industry's monthly magazine of applied electronic environering and development, Editorial, business, and market research staffs at booth to swap ideas with registrants. Free subscriptions at booth for engineers only.

Electronic Instrument Co., Inc. 33-00 Northern Blvd. Long Island City 4, N.Y. Booth 3505

Harry R. Ashley, Philip A. Portnoy, Mel Fink, Richard Dugot, Robert Clark Duff, Ike Rosenstein, Mannie Horowitz, Vince Proc, Kaz Komendowski



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Electronic test equipment, high fidelity components, ham equipment—kit or factory wired. VTVM's, oscilloscopes, tube testers, generators. VOM's, signal tracers, decade baxes, battery eliminators, amplifiers, preamplifiers, tuners, speakers, transmitters, modulators, grid dipmeter.

Electronic Measurements Co., Inc. Lewis St. and Maple Ave. Eatontown. N.J. Booths 2213-2215 Barney DeBlasio, John Raczek, Douglas K. Stevens, Sidney Norinsky, John Baugher, Paul Glasgow

Baugher, Paul Glasgow Regatran transistorized power supplies rated to 60 amperes. Regatron programmable power supplies for lab, remote control, or automatic systems application, liqh stability. Complete line of voltageregulated and constant-current power supplies. Precision calibrated power supplies.

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Morris Lieblich, Louis Stans, Jack Weber Tube Testers, Oscilloscopes, Signal Generators, Vacuum Tube Volumeters, Volt-Ohm-Milliammeters, Vibrator Checker, Battery Eliminators, Signal Tracers, Condenser Bridges, Stereo Amplifier.

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2444

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8

9

TOROIDAL FEATURES Reduces stray fields and proximity effects to obtain better stability. Permits small call construction Higher effective permeability Coupling not affected by tuning circuit High stability with temperature and time Low harmonic distortion Low harmonic distortion

Low narmonic distortion Improved insulation results in high Q Manufacturing methods permit close control of permeability and Q Finishes of tough thermosetting resins minimizes moisture absorption and provides insulation suitable for winding enameled wire directly on the core.

CORE SIZES

Cores are available in diameters from 9/32 OD to 2" OD Permeability: From 8 to 45 Recommended frequencies: Materials are available which will provide good Q from 0.1 to 25 MC

Write for samples and further information

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March, 1960

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TRANSISTORIZED **D-855** GAUSSMETER

- · Complete portability for use in field or lab
- Reads flux fields up to 30,000 gauss · Can be equipped to read Earth's
- field flux density
- Probe is only .025" thick
- Active area of probe .01 square inches
- Fully transistorized
- Power Supply: selective from 105-125 volt 50-60 cycle line or internal batteries
- Net weight: 8-3/4 lbs.
- Overall size: 13-1/2" high, 8-3/4" wide, 7-1/4" deep

Precision built, completely transistorized, the new D-855 Gaussmeter accurately measures flux density and de-termines "flow" direction. Ideal for measuring and locating "stray fields", plotting variations in strength and checking production lots against a standard. It's simple to operate. The Dyna D-855 doesn't require jerk or pull, gives no ballistic reading. Can be operated in the field with batteries which are enclosed in rugged protective carry case. This is an improved version of the pioneering D-79 Gaussmeter (Pat. #2,707,769) which has modernized magnetic flux measurement for the past six (6) years.



Whom and What to See at the Radio **Engineering Show**

(Continued from page 241.4)

Electro Motive Mfg. Co., Inc. South Park & John St. Willimantic, Conn. Booth 2731 J. Kevin Foley, Milton Lauter, A Arthur W. Evans, Joseph Regan, James Gilligan, Charles Rueb, Peter Nichols, Howard Ogushwitz, John Obsharsky, John Haines



Mylar-Paper Dipped Capacitors

Manufacturers of quality capacitors with built in reliability. Dipped Mica, Molded Mica, Sil-vered Mica Films, Dipped Mylar-Paper, Dipped Paper, Tubular Paper, Ceranic Discs, Ceranic Feed-Thrus, Variable Ceranics, Trimmers, Dual Padders and Single Padders.

Electro-Pulse, Inc. 11861 Teale St. Culver City, Calif. Booths 3810-3812 ▲ James S. Johnson, ▲ J. E. Niebuhr, Bernard Evans



5500A Variable Word Length Generator

Affiliated Servo Corp. of America, General Pur-Pulse Generators, Digital Pulse Generators, Current Generators and Core Testers, Time Mark Generators, Electronic Counters, Word Generators



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First and Second floors-Components Third floor—Instruments and Com-

plete Equipment Fourth floor-Materials, Services, Machinery

Electro Tec Corp. 10 Romanelli Ave. South Hackensack, N.J. Booth 1929

George J. Pandapas, Arthur Asch, Eu-gene Parks, P. J. Marshall, William Kennedy, Robert Kinsey, Robert Leslie, Herbert Gross, Peter Fillpczak, William Thacker, George Merker, Richard Koester

Miniature to giant Slip Ring Assemblies for instruments, gyros, & radar; Rotary Low Torque Switches; Mark II Relay for hi vibration; hi shock; hi temperature (200°C) and other hi-reliability applications.

Electro-Voice, Inc., Booth 1632 Cecil & Carroll Sts. Buchanan, Mich.

Everett Leedom Microphones, Cartridges.

Electrocraft, Inc., Booth 2126 See: G-C Electronics Mfg. Co.

Electrol, Inc., Booth 3940 9000 West Pico Blvd. Los Angeles 35, Calif.

R. A. Sinker, D. H. Sinker

Model 100 Diode Function Generator featuring a punched card memory, increased accuracy, greater computation capacity and ease of op-eration. DC amplifiers, chopper amplifiers and analog computers. analog computers.

Electron Corp., Div. Ling-Altec Electronics, Inc., Booth 3311 P. O. Box 5570

Dallas, Tex.

▲ Mort Zimmerman, Mike Ling, Sol Cornberg, Irving Presser

Complete Low-Power Television Broadcast Sta-Complete Low-Power relevision Broadcast Sta-tion Equipment, Amateur Television Station, Industrial & Educational TV Systems, TV Cameras, TV Monitors, TV Transmitters, Video & Audio Switches, Special Emphasis on Electron "Texan" TV Translator.

Electronic Applications, Inc., Booth 3929

194 Richmond Hill Ave. Stamford, Conn.

▲ Vincent J. Skee, Harry N. Reizes, John Cos-tello, Richard Wiggins, Truus M. Skee

tello, Richard Wiggins, Iruus M. Skee AKG Professional Microphones, EMT 325 Very Low Range Ohnneter 10 microolnns to 3000 Milliohns, EMT 543 Wide Range Electrolytic Direct Reading Bridge 100,000 Microfarad. Schomandl Precision Frequency Decades, FAPR F_x —300 Megacycles, Microwave Decade FD 3, Active Source Up To 12.6 Kilomegacycle.

Electronic Associates, Inc. Long Branch & Naberal Aves. Long Branch, N.J. Booths 3712-3718

R. L. Yeager, W. F. Blodgett, J. H. Smiley, J. D. Kennedy, B. Johnson, O. J. Sullivan, J. E. McCloskey, W. Peet, R. J. Doelger, W. O'Brien, A. T. Ashton



231R PACE Analog Computer

231R PACE Analog Computer, high speed re-petitive operation, digital programming equipment (ADIOS), X-Y analog and digital plotters (VARIPLOTTER, DATAFILOTTER), Desk-top analog computer (PACE TR-10), analog-digital and digital-analog computer linkage system (ADDALLNK), analog computer package for process control. (Continued on page 244.4)

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WHEN WRITING TO ADVERTISERS PLEASE MENTION-PROCEEDINGS OF THE IRE

March, 1960

(Continued from page 240.A)

Electro Devices Inc., Booth 4107 580 Main St.

Wilmington, Mass. Paul J. Post, Fred F. Cain, A. Allyn Ryalls, Edw. Gratto

Manufacturing Toroidal Coil Winders for Production (Model D6) and Laboratory (Model C). Core range 2" O.D. to 1/16" I.D. Wire 24AWG --46AWG--Speed 1000 Turns pm.

> Electro Instruments, Inc. 3540 Aero Dr. San Diego 11, Calif. Booths 3836-3838

▲ R. H. Applin, ▲ Joe Deavenport, B. Edelman, John Engelberger, Stanton East

All Electronic Analog-Digital and Digital-Analog Converters; Digital Voltmeters, Ratiometers, Ohummeters; Wilehand Transistorized Single-Ended or Differential DC Amplifiers; Solid State X-Y Recorders and Accessories; Transistorized Counters and Timers; Counter Systems; Data Acquisition Processing and Recording Systems.

Electro-Measurements, Inc., Booths 3010-3011

See: Electro Scientific Industries, Inc.

Electro-Mec Laboratory, Inc. 47-51 33rd St. Long Island City 1, N.Y. Booth 2312

▲ Forbes Morse, ▲ Robert Wiener, Bob Bordewick, ▲ Phil Luce, Carl Bernsten, Bob Ebert, George Boziwick, Gil Bassin, Dann Neubauer, Ellis C. Scovel, Jerry Bouton, Perc Ridley



Electro-Mec Type D3OU-10 DIGITOMETER ®

Potentionneters—Variable, Wirewound, Precision, Ultra Low Torque; Digitometers (Analog to Digital Converters); Goniometers (For Measuring and Testing of Potentiometers, Synchros and Similar Rotary Electronic Components); Servomechanisms, For Industrial and Military Aircraft Control Problems.

Electro-Mechanical Instrument Co., Booth 1231 8th and Chestnut Sts.

Perkasie, Pa.

L. R. Void, Ray Jones, R. I. Dinlocker, Robert Gombert

Ammeters, voltmeters and milliammeters, 2 inches to 4^{+}_{-1} inches. Instrument type miniature relays. Miniature flag type circuit indicators. Microammeters. Tuning meters.

(Continued on page 242A)

Be sure to see all four floors!

PROCEEDINGS OF THE IRE March, 1960

1. z = f(x,y) 2. z = f[g(x),h(y)] 3. $z = f(u \cdot x, v \cdot y)$ 4. $y_1 = f_1(x), y_2 = f_1(x), \dots -y_{20} = f_{20}(x)$ 5. $z_1 = f_1(x,y), \dots -z_4 = f_4(x,y)$ 6. $u = z \cdot f(x,y)$ 7. $z = f(x_1 + x_2 + \dots + y_1 + y_2 + \dots)$

NOW... A NEW APPROACH TO FUNCTION GENERATION

The Link Analog Function Generator

Link's analog function generator offers a new level of performance for analog computation and simulation. Key to this outstanding performance..., a Link-developed rectilinear servo motor with solid-state servo-amplifiers and a ceramic-film resistance element.

This new function generator eliminates the high drift and complex design of diodes generators, provides high-speed operation without the limited flexibility of optical techniques and the inherent backlash, friction and inertia problems of existing servo generators.

IT PROVIDES:

- **RELIABILITY** Modular design Automatic failure protection Simplified maintenance
 - ECONOMY-Standardized components Printed circuits
- FLEXIBILITY- Plug board programming Rack mounted or table top use
- **VERSATILITY** Numerous functions or function groups can be generated with minor modification, or by connecting one or more generators in series.

The analog function generator, first of a line of DIALOG* components and system building blocks to be introduced by Link, is another example of Link's unique computer capability. Thoroughly experienced in analog and digital techniques, Link can provide the most objective, economic solution to computation, simulation and control problems. For additional information on Link's new Function Generator or its broad computer capabilities – and your copy of Link's DIALOG* catalog – write to Industrial Sales Department.

DIALOG* (Link <u>Digital-Analog</u> System Components and Building Blocks)



*DIALOG is a Trademark of Link Division, General Precision, Inc.

BINGHAMTON, NEW YORK

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World Radio History

LINK DIVISION



(Continued from page 238A)

Electralab Printed **Electronics** Corp. Needham Heights 94, Mass. Booth 2130 ▲ Warren G. Abbott, ▲ Richard G. Zens, George McCarthy, Burt M. Isaac son, H. Eugene Jones Fined Wiring; Printed Circuit Assemblies; Cu-Con Plated Holes; Exclusively for High Reliability Electronics "Flugh Circuits with Cu-Con Plated Holes, "Protowiring Dept, Services Rush Re-quirements for Filot Runs, Protongaka, the "Do-it's burself" Laboratory Unit for Processing of Printed Wiring Boards. Electric Auto-Lite Co., Booth 4520 Champlain St. Toledo 1, Ohio S. T. Rose, V. R. Steinmeyer, S. W. Weiden-bach, R. Tweils, G. Rinebold Wire & Cable, Instruments, Ceramics, Electric Hotpack Co., Inc. 5019 Cottman Äve., Dept. #761 Philadelphia 35, Pa. Booths 3931-3933 Arnold Mann, Tom White, Ira McFarland, Bart Conchar, Douglas Bergen, Len Wingard, William Rariden, Morton Levy E ... D

Hot-Cold Temperature Test Chamber

Controlled Temperature & Humidity Chambers: Hot-Cold Test Chambers*. -100°F. to 400°F.; Walk-in Rooms, 0°C. to 40°C., ambient to 98% R.H.; Ovens, 35°C. to 100°C., 0°C. to 100°C. 20% to 98% R.H.; Vacuum Ovens, ambient to 200° and 300°C., vacuum to 1 Micron. For testing, conditioning and processing electronic components components.

Electrical Industries Division of Philips Electronics & Pharmaceutical Industries Corp. Murray Hill, N.J. Booths 2526-2528 O. H. Brewster, P. A. Muto, K. F. Mayers, J. Jonassen, D. Wilson, C. W. Beach Compression Type Threaded End Seals New expanded line of Compression Type Threaded End Seals, Glass-to-Metal Seals, Com-plete line standard teropinals, threaded seals, transistor and other miniature closures. Custom seals to specifications, and custom sealing of components.

(Continued on page 241.1)

Environmental Testing

A complete environmental testing service for Electronic, Electrical, Electro-mechanical, Mechanical, Hydraulic and Pneumatic components, products and systems to all Government and Industrial specifications.

Stressing the Engineering Approach in Environmental, Qualification and **Reliability** Testing

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March, 1960

TEL: EDGEWOOD 3-6650





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Oldsmobile Engineering Leadership sets the industry pace with a unique electronic wheel alignment device that dynamically computes "toe-in" measurements for precision steering and handling.

Handling and steering ease depend upon precise, minute measurement and control of front wheel alignment. Because wheels have a tendency to "toe-out" when in motion, they must be adjusted for a slight amount of "toe-in" to eliminate "wheel fight", wander and undue tire wear.

To meet the requirement of rapid, yet extremely accurate measurements on the production line, Oldsmobile engineers developed an electronic computor-a *linear*- *differential-variable transformer*—that dynamically and accurately measures the average amount of toe-in within .030 inches. As the car is brought into position, the wheels are rotated by rollers to simulate actual driving conditions and to eliminate errors caused by variations in tire run-out. By watching the visual gauges, an operator can quickly make the necessary adjustments to the steering linkage.

By using the most up-to-date electronic measuring techniques in engineering and manufacturing. Oldsmobile is able to offer safe, accurate steering and handling . . . a controlled, comfortable ride. Visit your local Oldsmobile Quality Dealer, take a ride in a '60 Oldsmobile and see why it's the value leader of its class!

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Nothing to add, nothing to fix. With the compact, powerful, rechargeable SIL-CAD you can forget about maintenance!

For this reason - and many others design engineers in military and commercial fields have turned to the rugged, lightweight, economical YARDNEY SIL-CAD to meet the demands of today's portable electronics - receivers, transmitters, computers, transverters, converters, solid state inverters, and all portable electronics...heavy-duty lighting, internal and ground support APU replacements, marine equipment, pagecall systems, stand-by power...portable medical equipment . . .



IRE SHOW BOOTH #2127

YARDNEY ELECTRIC CORP.

"Pioneers in Compact Power" R 40-50 LEONARD STREET, NEW YORK 13, NEW YORK Patents granted and pending. © 1960 by Yardney Electric Corp.

Whom and What to See at the Radio Engineering Show

(Continued from page 237A)

Ebauches S.A. (Neuchatel, Switzerland) c/o Freeport Engineering Co. 350 Fifth Ave. New York 1, N.Y. Booth 3947 Charles Friedman, A. Berger



Quartz Clock B-288

*New Model B-288 Quartz Clock Guaranteed Accuracy 1 \times 10⁻⁷ for frequency standards, time keeping, high precision chronometry, also precision transistorized impulse units (Decade, Binary and 10 KC Oscillator) for computer and general circuit application. Full engineering data available on these & other Ebanches Test Equipment from Laboratories, Observatories, etc.

Ecco Electronic Components Corp., Booth M-20 578 Nepperhan Ave.

Yonkers, N.Y. Leon Weiss, Walter E. Jezewski, Richard Swanson, John Morris, Malcolm Biener UHF Connectors, BNC Connectors, TNC Con-nectors, Type "N" and "C" Connectors, Tele-phone Plugs and Jacks, 125 Ohm Constant Im-pedance Connectors, Cable Assemblies.

Edgerton, Germeshausen & Grier, Inc., Booth 3914B

160 Brookline Ave. Boston 15, Mass.

M. D. Altfillisch, ▲ J. Murachver, ▲ R. L. Purrington, B. F. Roberts, W. A. Ward *Wide Band Traveling Wave Oscilloscope: *Fast Rise Pulse Generator; High Speed Oscilloscope Camera; *Ceramic Hydrogen Thyratrons; Nenon Flash Tubes and Driver.

Thomas A. Edison Industries Instrument Division 51 Lakeside Ave. West Orange, N.J. Booths 2739-2741

Victor W. Rose, Robert P. Zupa, George V. Boselli, Joseph C. Emma, John B. Norton, Arthur Douglas, Richard Thompson, Joseph Zercoe, Herbert F. Preiss, Edward J. Kerr



Model 323 R.F. Coaxial Cable

Servo motors, motor-generators and gearheads and electro-mechanical packages, time delay re-lays, thermostats and sensitive relays. Fire detec-tion systems, oil pressure transmitters and in-dicators for aircraft and spacecraft. High tem-perature 1000°F R.F. coaxial cable; has excel-lent attenuation and low weight; this 50 ohm cable also excellent for use under nuclear radia-tion. tion.

Eitel-McCullough, Inc. 301 Industrial Way San Carlos, Calif. Booths 2410-2112 Fred Johnstone, Hal Yokela, Cliff Warner, Bob Plummer, Al Melrose, Dave Wyand, Berk Baker



New developments in Eimac ceramic high power amplifier and reflex klystrons and negative grid tubes will be displayed. Klystrons include types extensively used in tropo-scatter applications.

Elastic Stop Nut Corp. of America. Booth 2343 See: A'G'A Division.

> Eleo Corp. "M" St. below Erie Ave. Philadelphia 29, Pa. Booth 1515

▲ Benjamin Fox, Leo Kagan, ▲ H. E. Ruehle-mann, ▲ Samuel Weiss



Series 8105 Varicon Connector

Series 8007-8008 Rack-and-Panel or Cable Con-nectors. Series 8007, with screw actuation: 75, 100, 130 contacts. Series 8008, without screw actuation: 80, 95, 110, 125, 140 contacts. Both Series available with or without cover. Feature Varicon contacts. *New micro-miniature con-Varicon contacts, nector Series,

Eldico Electronics Div., Booths 1202, 1301-1303

See: Radio Engineering Laboratories, Inc.

Electra Manufacturing Co., Booths 2834-2836 4051 Broadway

Kansas City 11, Mo.

Gordon Groth, W. E. McLean, R. T. Means, A. B. Mayer, R. R. Burton R. B. Mayer, R. R. Sensor, Deposited Carbon, Metal Film and High Temperature, as well as a complete line of Ceramic Disc and Plate Ca-

pacitors. (Continued on page 240A)

Your firm not listed?

If you would like to become an exhibitor in the 1961 show, write for information to IRE Exhibits Dept., 72 W. 45th St., New York 36.

E.M.I.-Cossor Electronics Ltd. 2005 McKay St. Montreal, Canada Booth 1520

▲ Howard L. Hoffman, ▲ Robert Sheviot, Arthur Hoffman, Fred Belasco, Aldo Pulvirenti, ▲ C. M. Morgan, ▲ J. Sharpe, H. Clarke



Complete line of *Klystrons, *Photomultipliers, *Traveling Wave Tubes, Orthicons, Vidicons, Magnetrons, Secondary Emission Tubes, Special Purpose Tubes, Cathode Ray Tubes, *Wide Band Oscilloscopes, Waveguide Test Equipment, *Sanders Microwave Instruments.

E.M.I. Electronics Ltd., Booth 1520 See: E.M.I. Cossor Electronics Ltd.

ESC Corporation 534 Bergen Blvd. Palisades Park, N.J. Booth 2915 Morton Fassberg, A Stanley S. Packer, Bernard Brain, A Rod Yard, A Stanley Pearl



Pushbutton Decade Delay Unit

Video Delay Lines, fixed, variable, lumped and distributed, of all types; Pulse Transformers, Wide Band Video Transformers, Filters, Toroids and other networks.



(Continued on page 238A)

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A complete listing of all registrants at the IRE National Convention and Radio Engineering Show, showing company affiliation and local hotel, is posted on the back wall of the mezzanine on the west side of the first floor.

For Economy AND Quality....





(Continued from page 235A)

EH Research Laboratories, Inc. 1922 Park Blvd. Oakland 6, Calif. Booth 3025 ▲ William M. Brobeck, Dr. John C. Hubbs, C. E. Andressen, Jr., William D. Jordan D. Jordan Fast rise time high rep rate Pulse Gen-erators, Transistor Rise-Storage-Fall Time Meter for measuring switching transient characteristics. Electrometers for measuring currents to 10⁻⁴⁵ ampere, Multi-Channel Coincidence Units, Milli-voltmeters, Megmegohumeters, Milli-voltmeters, Megmegohumeters, Milli-croammeters, Current Integrators.

EICO. Booth 3505 See: Electronic Instrument Co., Inc.

▲ Indicates TRE member. Indicates new product

ELIN Division International Electronic **Research** Corporation 145 West Magnolia Blvd. Burbank, Calif. Booth 3018 Harvey Riggs, Don Rich, Fred Miller, ▲ John E. Markley, Jr., John R. Foster, Otto Leichliter



Three-Phase Power Supply

Complete Line of Regulated AC Power Sources. The New *3-Phase Power Supply with 0.1% Regulation, 0.2% Maximum Distortion and a Power Load Factor Up to 0.3 Inductive Will Be Shown for the First Time Along with Our New *Voltage Calibrator and Complete Line of Single Phase Precision Power Oscillators.

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(Continued from page 234A)

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(Continued on page 236.4)

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(Continued from page 233.1)

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(Continued on page 235A)

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> Wilbur B. Driver Co. 1875 McCarter Hgwy. Newark 4, N.J.

Booths 4201-4203

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(Continued on base 234A)

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Filament transformer for insulation up to 80 KV AC Test. Low secondary capacitance from 6 to 30 mmfd.



High impedance type transformer from 0.01 to 50 KVA and up to 10 KV. This unit is used for applications where short circuit current must be limited



This transformer features low voltage high current secondary windings up to 4000 amps., and up to 300 KVA. Taps on the primary windings afford a wide range secondary current.



Same as opposite except with 2 or more secondary windings.



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Whom and What to See at the Radio Engineering Show

(Continued from page 231A)

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Diehl Manufacturing Co., Booths 1819-1823

See: Singer Manufacturing Co.

▲ Indicates 1RE member. ■ Indicates new product. Digital Equipment Corporation Maynard, Mass. Booth 3831 A Kenneth H. Olsen, A Harlan E. Anderson, A Richard L. Best, John B. Brown, A Jonathan Fadiman, A Benjamin M. Gurley, Robert A. Hughes, Stanley C. Olsen, A Walter E. Weeton



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232A

March, 1960

(Continued from page 226.4)

Deutsch Company, Electronic Components Div., Booth 1425 Banning, Calif.

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See: Cornell-Dubilier Electric Corp.

Dialight Corp. 60 Stewart Ave. Brooklyn 37, N.Y. Booths 2829-2831 ▲ R. E. Greene, M. Greene, H. W. G ▲ J. L. Weil, M. Roberts, R. B. King Goodman,



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Antennas, Multi Channel Rotary Joints*, Hi-Power Waveguide Rotary Joints*, High Power Coaxial Switches*, High Power RF Dummy Loads (Air Cooled)*, Low Loss Attenuators*, Diplexers, Directional Couplers, Frequency Meters, Hybrids, Phase Shifters, Waveguide & Coaxial Assemblies, Slotted Lines, Accessories.

(Continued on page 232A)

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Check your requirements against these SAGE "Silicohm" Resistors

TYPE "NS" SILICONE COATED RESISTORS



Sage "Silicohm" Type "NS" units are compact, light weight. Exclusive insulation for extreme combinations of moisture and temperature environment (-65°C to 350°C) . . . dielectric strength-1000 volts RMS . . . precision to .05% + 20 ~ /or

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Style	ratings, WATTS	dime L	nsions D
NSA2W	2	.500	.187
NSB2W	2	.812	.187
NS2W	2	625	.250
NS3W	3	.750	.250
NSS5W	5	.875	.312
NSR5W	5	1.000	.312
NSL5W	5	1.125	.312
NSS7W	7	1.250	.312
NSR7W	8	1.375	.375
NSS10W	10	1.812	.375
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TYPE "NM" METAL CLAD CHASSIS MOUNTED RESISTORS

Aluminum housed resistors for heat sink mounting, Type "NM" units feature considerably less heat rise than any other resistors of comparable size and wattage.



Lower hot spot means longer service life, near perfect stability-(Average resistance shift is only 0.4% after 1000 cycled hours at recommended loads), and exceptional reliability under extreme conditions. Dielectric strength: 1000 volts RMS to 2500 volts RMS, equal or exceeding Mil requirements . . . precision to .05%.

Style	ratings, WATTS	Nominal Mounting Dimensions, inches A B		
NM10W	10	.562	.625	
NM25W	20	.719	.781	
NM50W	40	1.562	.844	

Write for samples and complete specifications.



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NEW HI-FI TYPE **SYLVANIA 7687** CONTROLLED FOR LOW HUM

The new 7687 is a 9-pin miniature triode-pentode controlled for hum, noise and microphonics. It's a hard worker in tone-control amplifiers, phase splitter and high-gain voltage amplifier circuits, yet it does its job without even "breathing audibly." Sylvania 7687 structure is rigidly mounted to reduce noise and microphonic effects. It features a cooler-operating cathode to assure low hum. Further assurance of low hum is provided by the use of a coil heater made of specially developed materials. The triode section has an equivalent hum and noise level of 7.5 microvolts, the pentode only 10.5 microvolts. Investigate the possibilities of a cooler-operating tube with unusually low hum and long life expectancy for your compact high-fidelity design. The Sylvania 7687 merits your interest.

SYLVANIA "GLEAM" PROJECT COMBATS TUBE CONTAMINANTS, INCREASES TUBE RELIABILITY

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World Radio History

SYLVANIA ANNOUNCES **3 NEW TUBE TYPES** WITH 9-T9 OUTLINE!

New 17HC8, 6HC8 and 7695 offer important advantages inherent in the Sylvania unique 9-T9 design. Utilizing the straight-sided, 9-T9 bantam outline with its miniature 9-pin circle, these three types afford significant opportunities for compactness. The 9-T9 outline eliminates the octal base of the T9 and makes possible the use of tube structures capable of high plate dissipation in printedcircuit boards. This is accomplished with conventional 9-pin sockets widely used in printed circuits.

9-T9 increases volumetric efficiency of the chassis by eliminating the octal base of the T9 outline.

9-T9 enables the use of large tube-assemblies in those stages where higher power-dissipation capabilities of the tube are a design necessity to enhance reliability.

9-T9 maintains compactness of the equipment formerly afforded by tubes fitted with T6- $\frac{1}{2}$ header.

Sylvania 17HC8 is a triode-pentode designed for use as a vertical deflection oscillator and vertical deflection amplifier in 110° deflection circuits of TV receivers. Controlled for heater warm-up time, it is especially useful in 450mA series string operation. The pentode section has a plate dissipation of 11 watts. Structure of the 17HC8 includes an internal shield to reduce interaction of the ele-



ments. The 6HC8 is identical to the 17HC8 except for heater power requirements. In addition to normal 100% tests for shorts, continuity, plate current, gas, pentode screen current, heater cathode leakage, gm and triode cutoff, both types are tested 100% for peak plate and screen current, ratio of peak plate current to screen current, and microphonics.

Sylvania 7695, beam power pentode, features remarkably high power sensitivity as an audio frequency amplifier. In Class A1 operation, it can deliver 4.5 watts of power with a B+ voltage of only 130 volts. As a result, the 7695 makes possible economies in power supply requirements.

SYLVANIA 7695-Characteristics and Typical Operation				
Fixed Bias	Self Bias			
	140 Volts			
130 Volts				
—11 Volts				
_				
11 Volts	8 Volts			
95 mA				
100 mA				
5 mA	5 mA			
6.900 Ohms	6.800 Ohms			
1.100 Ohms	1,100 Ohms			
	4.5			
11 Percent	11 Percent			
	Stics and lypica Fixed Bias 130 Volts 130 Volts			





World Radio History

NEW – PICTURE TUBES WITH ELECTRONIC SCAN-MAGNIFICATION!

SYLVANIA ST-2836A — now in the developmental stages — incorporates a mesh-like diverging-lens assembly positioned in the neck of the tube. Its function is to provide deflection of the electron beam in addition to that accomplished by the magnetic field of the yoke assembly. The linear magnification of scan is in the order of two tlmes for an anode-to-mesh

voltage ratio of 2 to 1. The primary benefit of such a technique is in the reduction of horizontal-deflection power requirements. It is anticipated that this power requirement may be reduced in practice to as much as 60% of that required for conventional 110° picture tubes. Engineering samples with *lowpower heaters* (1.5-volts @ 140 ma., or 12.6-volts @ 150 ma.) and/or low Eg₂ characteristics for a *complete* low-power picture tube are also available. For technical data and further information on SYLVANIA experimental-design SCAN-MAGNI-FIED PICTURE TUBES, contact the Sylvania Field Office nearest you. BUE BUE SCAN MACHINER TUBE RAM DEVICENCE WITHOUT LINE ACTION DEVICENCE WITHOUT LINE ACTION DEVICENCE WITHOUT LINE ACTION DEVICENCE ACTION DUE TO LINE ARD

NEW-HIGH-VISIBILITY 'SCOPE TUBE FOR AIRBORNE WEATHER RADARI

SYLVANIA SC-2854 provides improved image brilliance under wide ambient light conditions encountered in cockpits of commercial airliners. The color of the phosphor of this new tube gives exceptional image visibility to dark-adapted as well as to light-adapted eyes. Resolution, too, is exceptionally high. *Sylvania SC-2854* makes possible simplified equipment designs, improved volumetric efficiency and increased life-expectancy of the indicator tube, resulting in reduced costs of installation and maintenance of airborne weather-radar equipment. For details on price and delivery, contact your Sylvania Field Office.

NEW – C.R.T.'S FOR HIGH-ALTITUDE OPERATION TO 70,000 FEET!

Sylvania now makes available a group of direct-view cathode-ray tubes designed specifically for applications in airborne ECM, Radar, and Loran equipment intended for operation at high altitudes. All types feature high quality, nearly flat pressed-glass faceplates. This provides exceptionally clear display and excellent bulb strength. Connections to internal elements are made through insulated leads, encapsulated at points of entry to the bulb. This technique significantly reduces the possibility of corona and arc-over at high altitudes. See data below.

SYLVANIA SCVP1, SCVP7, SCVP19... feature 2¾" x 4¾" directview faces, magnetic deflection, electrostatic focus.

MAXIMUM RATINGS (Absolute Maximum Values)

Anode Voltage	O Volts dc
Anode Input	6 Watts
Grid No. 4 Voltage (Focusing Electrode) 500 to +110	0 Volts dc
Grid No. 2 Voltage	0 Volts dc
Grid No. 1 Voltage	
Negative Bias Value 16	5 Volts dc
Positive Bias Value	0 Volts dc
Positive Peak Value	2 Volts
Peak Heater-Cathode Voltage	
Heater Negative with Respect to Cathode 18	0 Voits
Heater Positive with Respect to Cathode 18	0 Volts
Altitude	0 Feet
Operating Temperature Range	0



SYLVANIA 3BEP1, 3BEP-... feature 1½" x 3" direct-view faces, electrostatic focus and electrostatic deflection. (-* can be supplied with several other screen phosphors.)

MAXIMUM RATING5 (Absolute Maximum Values)

Anode No. 2 Voltage	000	Volts dc
Anode No. 1 Voltage (Focusing Electrode)1	200	Volts dc
Grid No. 1 Voltage		
Negative Bias Value	140	Volts dc
Positive Bias Value	0	Volts dc
Positive Peak Value	2	Volts
Peak Heater-Cathode Voltage		
Heater Negative with Respect to Cathode	140	Volts
Heater Positive with Respect to Cathode	140	Volts
Aititude	000	Feet
Operating Temperature Range65 to +8	5°C	

Electron Tube Newsfrom SYLVANIA

TV PICTURE IS "UP FRONT" .SALES ARE, TOO

...when you design around Sylvania 23" and 19" "Bonded Shield" TV picture tubes!

SYLVANIA pioneered the techniques that make possible the quantity production of the new "Bonded Shield" picture tubes for TV sets. SYLVANIA led the way by making "Bonded Shield" picture tubes available to TV set manufacturers in commercial quantities. SYLVANIA was first to demonstrate how "Bonded Shield" eliminates the "picture-in-a-tunnel" effects; first to demonstrate the possibilities of "broad-angle viewing" dramatically offered by this new design.

An annealed-glass scratch-resistant cap is laminated to the face of the tube. It completely eliminates the need for a front-of-the-cabinet safety glass. This reduces reflections that interfere with the brilliance and clarity of the TV picture. Further, it reduces basic requirements for front-to-back dimensions of the TV cabinet, creating new possibilities for cabinet styling and sales appeal. The laminated safety cap eliminates the dust trap between tube face and safety glass. Corners are squared to give larger picture areas. Integral safety-glass and mounting lugs add up to potential savings in costs of cabinetry. Now, "Bonded Shield" picture tubes are also available with non-glare coating. They offer freedom from undesirable reflections and glare.

For technical data and further information, contact the Sylvania Field Office nearest you. Reduces Dangers of Implosion

- Minimizes Production-Line Rejects
- Simplifies Mounting
- Reduces Reflection up to 50%
- Squared-Corner Screen
- Offers New Cabinet Design Flexibility

SYLVANIA 23's

282 sq. in.

viewing area!

SYLVANIA 19's

viewing area!

- 178 sg. in.

"CLOVERLEAF" Ceramic Cathode Assembly in every "BONDED SHIELD" Picture Tube!

... assures fast warmup time throughout tube life. Sylvania developed this unique structure to reduce heat conduction losses and to give increased durability to the cathode assembly, resulting in improvements in tube life expectancy. For full details on the SYLVANIA "CLOVER-LEAF" and its benefits, contact the Sylvania Field Office nearest you.
HIGH ENSI



micro-miniature

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- Contiguous flush-mount
- 47-10,000 mmf
- 200 vdc without derating
- −55°C to 150°C operation

"VK" capacitors are designed with square precision molded cases in only two sizes and a single standard 0.2" lead spacing for all values. Continuous life and environmental testing, plus 100% tests for Dissipation Factor, Insulation Resistance, and Capacitance guarantee that each "VK" capacitor in your circuit will perform as predicted.

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Same electrical characteristics as stand-ard "VK" series. Each unit coated with a resilient protective compound. Dimen-sions: 47-100 mmf, .100" square; 120-270 mmf, .130" square; 330-1000 mmf, .150" square; 1200-3300 mmf, .250" square; 3900-10,000 mmf, .265" square.



Whom and What to See at the Radio **Engineering Show**

(Continued from page 225A)

DeJur-Amsco Corporation 45-01 Northern Blvd. Long Island City 1, N.Y. Booths 2307-2309 N. J. Geldman, P. J. Morrisey, E. Redgate, S. Dexter, L. Callan, V. Stein, R. Heitz, E. Drewitz, L. Zielinski, E. Brautigam, S. Gio-vino, J. O'Brien, D. A. Harkavy, J. Grohowski



2" Ball-Bearing Precision Potentiometer

Precision linear and non-linear, single turn potentiometers from $\frac{1}{2}$ " to 5", special $\frac{1}{2}$ " knob potentiometer*, custom panel instruments includ-ing VU and DB, aircraft flight instrument, rug-gedized types from $\frac{1}{2}$." to $\frac{3}{2}$ ", elapsed time in-dicators, and precision electrical Continental Con-peters

Del Electronics Corp. 521 Homestead Ave. Mount Vernon, N.Y. Booth 1816 J. G. Delcau, H. J. Di Giovanni, R. Kaufman, D. R. Congiusti, ▲ S. Glassman, I. Brill



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▲ Indicates IRE member. * Indicates new product.

Deltime, Inc., Booth 1725 139 Hoyt St. Mamaroneck, N.Y.

Mille Stand, Casper M. Bower, George Hoose, Jess Silberstein, ▲ Albert E. Powell, Thomas Dundon

Magnetostrictive delay lines, standard fixed and variable models. New developments in smaller units with higher storage capacity.

Deluxe Coils, Inc. P.O. Box 318 Wabash, Ind. Booth 1930

Ronald W. Keipper, James E. Bu Lewis Dumbauld, George I. Martin Brumbaugh.



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DeMornay-Bonardi 780 S. Arroyo Pkwy. Pasadena, Calif. Booths 3216-3218 Louis Della Penna, William T. Brock



* DBW-715 Cavity Wavemeter, 90-140 KMc

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Dempa Shinbun, Inc., Booth 2935 See: Japan Electric.

Derivation & Tabulation Associates, Inc., Booth 4113 95 Harrison Ave. West Orange, N.J. ▲ Henry Tulchin, E. L. Ayres

Characteristics Tabulations: Transistors, Semi-conductor Dioles & Rectifiers, Microwave Tubes, Updated Semiannually.

Design Tool Corporation, Booth 4122 772 Bergen St.

Brooklyn 38, N.Y.

Brooklyn 38, N.Y. Carl Kertesz, Wallace Krakauer, John Thomp-sen, Dudley Bell, Irving Kertesz *Automatic Spaghetti Attachment. Auto-Former Combination cuts insulation tubing from reels, inserts resistor (etc.) leads in tubing. Leads cut, formed, adjustable—Panto-Sert inserts com-ponents over entire printed board—Transistor Lead Fabricator—Axial Lead Straightener and Taper—*Economy Bender—Auto-Solder.

(Continued on page 231A)

Danbury-Knudsen Division, Booths 2402-2408, 2501-2507 See: Amphenol-Borg Electronics Corp.

Data-Stor Div., Booth 3110 See: Cook Electric Co.

Datex Corporation, Subsid. Giannini Controls Corp., Booth 3234 1307 S. Myrtle Ave. Monrovia, Calif.

J. L. Kent

Shaft position encoders, V-scan encoders*, Mini-module chassis*, precision encoder-servo assem-bly, Gray-to-binary translator*, and other com-ponents.

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Resistors.

Davies Division, Booth 2214 See: Minneapolis-Honeywell,

Daystrom, Incorporated, Booth 1807 430 Mountain Ave. Murray Hill, N.J.

Ann Mitchell Electronic Test Equipment, Panel Instruments, Portable Instruments, Relays, Resistors, Gyros, Potentiometers, Hi-Fidelity Equipment and Test Instruments.

Daystrom, Incorporated, Heath Company, Booths 1702 & 1801-1803 See: Heath Company, Daystrom, Incorporated.

Daystrom, Incorporated, Pacific Division, Booths 1704-1706 See: Pacific Division, Daystrom, Incorporated.

Daystrom, Incorporated, Transicoil Division, Booth 1805

See: Transicoil Division, Daystrom, Incorporated,

Daystrom, Incorporated, Weston Instruments Div., Booths 1708-1710 & 1809 See: Weston Instruments Div., Daystrom, Incorporated.

Decade Instruments Co., Booths 3512-3518

See: Kay Electric Co.

(Continued on page 226A)

▲ Indicates IRE member. Indicates new product.

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Whom and What to See at the Radio Engineering Show

(Continued from tage 222.4)



Cutler-Hammer, Inc., Booths 3316-3318, 3413-3417

See: Airborne Instruments Lab. Division.

D & R, Ltd., Booth M-24 402 East Gutierrez St. Santa Barbara, Calif.

B. C. "Buck" Rogers, Bob Lindbery, ▲ Ray L. Dawiey

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Dage Electric Co., Inc. 67 N. 2nd St. Beech Grove, Ind.

Booth 2433

A. N. Strickland, M. H. Burdett, W. O. Slater

Dage Electric will be exhibiting coaxial cable connectors, triaxial cable connectors and glass-to-metal hermetic scals.

Dage Television Div., Booths 1431-1631 See: Thompson Ramo Wooldridge, Inc.

> Date Products, Inc. 1302 28th Ave. Columbus, Neb. Booths 2627-2629

I. E. Gates, Dan Geeding, Jim Brandfas, Ray Root, J. Matejka

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Whom and What to See at the Radio Engineering Show

(Continued from page 221A)

Corning Glass Works Corning Electronic Components Bradford, Pa. Booths 2334-2336, 2427-2429

A. W. Dawson, C. C. Harword, J. F. Riley, M. R. Berell, B. D. Roesch, R. V. Hamjian, H. W. Hanson, J. G. Landers, U. H. Martz, C. J. Lucy, L. S. Moshier, J. G. Curtis, A. J. Hotte, L. S. King, K. S. McIntosh, N. Lazar, R. J. Settzo



Fotoceram .10" grid board

Tin-oxide resistors, fixed glass capacitors, FOTOCERAM printed circuit boards, ultrasonic delay lines, trimmer capacitors, rectifier tubes, metallized bushings, delay line coil forms, metallized glass inductors.

Cowan Publishing Corp., Booth 4215 300 W. 43rd St. New York 36, N.Y.

Sanford R. Cowan, Richard A. Cowan, Harold Weisner, Cary C. Cowan, Jack N. Schneider Semiconductor Products Magazine, CQ Maga-zine, Technical Books, Technical Publishing zine, Te Services.

Craig Systems, Inc., Booth 1325 360 Merrimack St. Lawrence, Mass.

Michael J. Macdonald, Stephen M. Friedrich, ▲ J. Roy Wolfskill, Gil A. Barrett, ▲ Walter White, Jr., Bernard C. Victory

Mobile electronic ground support equipment for communication, navigation and missile systems including air transportable shelters, trailer vans, control towers, missile carriers, spare parts con-tainers, relay racks and *portable telescoping antenna masts.

Crescent Petroleum Corp., Booth 2134 See: Kurman Electric Co.

Crosby-Teletronics Corp., Booths 3312-3314

54 Hinkel St.

Westbury, L.I., N.Y. ▲ Bart Coffman, ▲ Bob Constable, Bob Corbey, ▲ Ted Nelson, ▲ John Peters, ▲ John Sim-mons, ▲ Hank Schweibert, Frank White

Single-sideband receivers, transmitters, signal generators, Facsimile transmitting and receiving converters. FM Stereo Multiplex equipment. Pulse generators, counters and timers, Tele-metering oscillators, Diode testers, Silicon testers, etc

Crosley Division, Booth 3064 See: Avco Corp., Crosley Div.

▲ Indicates IRE member. * Indicates new product.

Crucible Steel Company of America, Booth 1327

P.O. Box 2518

Pittsburgh 30, Pa.

Pittsburgh 30, Pa. W. G. Scharnberger, J. R. Hansen, J. A. Byrnes, A.E. M. Underhill, Irene Wagner, J. A. Stavrolakis, C. A. Julian, E. F. Anderson, L. Benjamin, Frank Brinkerhoff, Jerry Carman, N. J. Carpenter, F. E. Chepko, M. DeChristo-pher, A. J. DeCosta, Joseph A. Driscoll, B. Dun-net, W. E. Gardner, George Hamamjian, J. Hennessey, C. A. Hirschi, David Hume, J. R. Knox, E. L'Esperance, J. Lucy, George Lyon, Arthur Manger, R. J. Patrick, George M. Red-gate, J. Sharkey, D. H. Sheridan, J. G. Thomas, I. S. Warren, T. D'Amico, J. S. Davis, J. Dougherty, D. F. Hall, A. Heath, H. Hughes, Arthur Kluglein, L. Kranes, P. J. McConnell, F. J. McNiff, E. G. Malan, J. L. Martin, R. Martorelli, J. R. Millikin, H. Rohner, J. Schmidt, W. Whaley

Permanent Magnets and Magnetic Component Assemblies,

Cubic Corp. 5575 Kearny Villa Rd. San Diego 11, Calif. Booth 3235

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James Cunningham, Son & Co., Inc. P.O. Box 516 33 Litchfield St. Rochester 8, N.Y. Booth 2237

Booth 2237 P. F. Cunningham, A Fred Bartlett, A John DeWolf, A.A. W. Vincent, R. P. Kennedy, A Frank Bradley, A Benjamin Margolin, Norman Silber-berg, A Robert Kelly, Tom Jones, A.R. Edward Stemm, Howard Carlson, Victor Sykes, A Jim Luscombe, A William H. Bradley, Myron R. Smith, Edward E. Bock, George R. Appleton, Allan Thaw, A. J. Vick, Emory Lane, Bud McDonald, Herbert Gentry, Herb Bass Complete line of Crossbar Switches for data handling, scanning, monitoring, tele-metering. testing equipment, automatic control, television broadcast, radar, sonar, thermocouple and strain gauge switching. "Self-stepping Crossbar Scanner. Also switching and scanning systems. High speed miniature Solenoid Actuators.

Curtis Development & Mfg. Co., Booth 2313

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(Continued on page 224A)

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(Continued from page 220.4)

Controls Company of America, Booths 1727, 1727A, 1727B See: Hetherington Switch Div. & Electrosnap Switch Div.

Cook Electric Co. Data-Stor Div. 8100 Monticello Ave. Skokie, III. Booth 3110 E. L. Washburn, A S. Himmelstein, R. S. Tveter, F. P. McGowan, R. Parks, R. Wahrer, A H. Grimme Model 59 all solid state, high speed digital magnetic recording system and Model 59 photoelectric reader. Model DR-25-2 solid state airborne magnetic recording system. Model 750-8309 solid state militarized high speed photoelectric reader. Model 750-7300 solid state militarized high speed digital magnetic recorder.

Coors Porcelain Co., Booths 4005-4006 600 Ninth St. Golden, Colo.

▲ Joe Coors, R. Schulze, W. G. McDonald, J. McManus, L. C. Hageman, Dan Howes High alumina components, Ceramic to metal assemblies—custom and standard. Beo components and Assemblies. Standard high alumina terminals and hermetic seal assemblies.

Cornell-Dubilier Electric Corp. Affiliated Federal Pacific Electric Co. 333 Hamilton Blvd. South Plainfield, N.J. Booths 2725-2727 R. T. Leary, A. Williams, G. E. Ronk, R. J. Reigel, A. Loeffler, J. Feder, J. Glynn, J. Cox, Lou Alexander, Hy Steinberg



Capacitors, Semiconductors, Vibrators, Power Supplies, Antenna Rotors, Delay Lines*, Wave & Noise Filters*, Pulse Networks*, Energy Storage Capacitors*, Relays*.

Corning Glass Works Corning, N.Y.

Booths 2334-2336, 2427-2429 L. A. Amylon, G. T. Backer, H. E. Bahr, E. J. Collins, H. S. Craumer, J. O. Cumiskey, J. S. DeMaio, R. H. Hildebrand, C. Howe, W. H. Hudson, R. L. Jones, E. C. Kramer, P. C. Leffel, V. B. Level, W. Linn, D. MacMillan, J. C. Marx, S. H. McKibben, T. G. O'Leary, F. E. Rector, P. L. Roederer, M. F. Rogers, M. R. Shaw, H. R. Silbaugh, J. L. Webb, R. K. Whitney

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(Continued on page 222A)

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Whom and What to See at the Radio **Engineering Show**

(Continued from page 219A)

Continental-Diamond Fibre Corp., Subsid. The Budd Company, Booths 4224-4226

Newark 48, Delaware O. Thomas, W. Jahns, H. Plate, H. Howe, A. Pronchick, A. Buck, D. Sullivan

Florence, A. Buck, D. Sunivan *Epoxy glass & paper base grades. Flame re-tardant grades. *Di-Clad printed circuit material. Products of Teflon, Silicones, and Polyester glass. Plus an extensive line of electrical insula-tion made from Diamond vulcanized fibre, Di-lecto, Celoron, Vulcoid, and Micabond, Fabri-cated north lecto, Celor cated parts.

Control, A Division of Magnetics, Inc. Box 391 Butler, Pa Booth 2437 Arthur O. Black, & Robert W. Olmsted, James W. Graham, H. B. Opitz, N. Altman, H. A. Savisky, W. S. Spring, J. E. Frauenheim, W. J. Irvine, R. C. Woodward



Standard Power Control Unit

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Control Electronics Co., Inc. 10 Stepar PL Huntington Station, L.I., N.Y. Booth 1902

▲ Alfred C, Walker, Eugene S. Wendolkowski, ▲ Marcus M. Epstein, Kieran R. Dunne, Hun-ter McShan, Vincent Pirro, Frank Battista



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Control Switch Div., Controls Co. of America, Booths 1727, 1727A, 1727B See: Hetherington Switch Div & Electrosnap Switch Div

(Continued on tage 221A)

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IRE SHOW BOOTHS 2713, 2715

(Continued from page 218.4)

Consolidated Diesel Electric Corp., Booth 1728B

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Stamford, Conn.

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Consolidated Mining & Smelting Co. of Canada, Ltd. 215 St. James St., West Montreal, Canada Booth 4511

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Consolidated Resistance Co. of America, Inc., Booth 1100A 44 Prospect St.

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(Continued on page 220.4)



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See Con Avionics at the IRE Show, Booth 1728A



(Continued from page 216A)

Computer Systems, Inc. 611 Broadway New York 12, N.Y. Booths 3837-3839 Robert K. Stern, Charles B. Husick, W. George Van Vliet, Salvatore J. Teta, Jack M. Andrews, Irwin West, B. Brachman, M. Schwartz



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(Continued on page 219.4)

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(Continued from page 215A)

Columbia Broadcasting System, Inc., Booths 1208-1210 & 3063 See: CBS Electronics Div. & CBS Laboratories Div.

Columbia Technical Corp. 61-02 31st Ave. Woodside 77, N.Y. Booth 1112 V. Liebmann, ▲ J. Machill, ▲ D. R. Stein, G. H. Weiland, N. Ordjanian



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(Continued on page 218A)

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(Continued on page 216A)

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(Continued from page 213A)

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Chilton Publishing Co., Booths 4301-4303

See: Electronic Industries

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214A

(Continued from page 212.4)

Chassis-Trak, Inc. 525 S. Webster Ave. Indianapolis 19, Ind. Booth 4001

▲ Larry M. Vaughn, Sarah Gray, Lou Flagin, John McShay



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Chatham Electronics Div., Booths 2428-2430, 2521-2523

See: Tung-Sol Electric, Inc.

Chemo Products, Inc., Booth 4056 100 Pulaski St., P.O. Box 169 West Warwick, R.I.

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(Continued on page 214.4)







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(Continued from page 210A)

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a Garth E. Bower, Sigmund P. Rosen, Daniel R. Kursman, Jerome Kursman, ▲ Ralph Weinger, ▲ Winston Starks, Bob Cameron, Paul Stock, Bernie Spector

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(Continued on Jage 213.4)





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Cannon Electric Company 3208 Humboldt St. Los Angeles 31, Calif. Booths 2628-2632

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(Continued on page 210A)

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Whom and What to See at the Radio Engineering Show

(Continued from page 206A)

CBS Electronics Div., Columbia Broadcasting System, Inc. 100 Endicott St. Danvers, Mass. Booths 1208-1210 Q. Adams, & W. Bevitt, & E. Boise, & R. Crosby, & J. Cunningham, C. Dibling, & R. Gibson, & J. Shenk, & R. Swain, R. Tomer, G. Wilde, & R. Yeiter



CBS Electronics will display microelectronics, 0.015% AQL indium-bonded diodes, NPN switching transistors, complementary NPN and PNP power transistors and frame-grid tubes, Also exhibited will be entertainment receiving tubes, audio components, secondary-emission tubes, krytrons, and other specialized tubes for instrument applications.

▲ Indicates TRE member, * Indicates new product, CBS Laboratories, Div. Columbia Broadcasting System, Inc., Booth 3063 High Ridge Road Stamford, Conn. David Alan Safer Vidiac Generator, Vac Bearings Systems.

CGS Laboratories, Inc., Booths 3803-3805

Wilton, Conn.

▲ M. L. Jackson, W. L. Gustavson, A. Winter, J. L. Gray

J. L. Gray *Antenna multicoupler for 2-32 mc; *50-100 mc panoramic receiver; *miniature INCREDUC-TOR high frequency saturable reactors; *tunnel diode characteristic tester; Morse to teleprinter code converter.

C & K Components, Inc., Booth 1627 101 Morse St. Newton 58, Mass.

▲ Robert H. Sturdy, ▲ Charles M. Sutherland, Marshall M. Kincaid, ▲ Charles A. Coolidge, David E. Miller, Robert Clonan, Richard H. Shute, ▲ Franklin Hobbs

Encapsulated magnetic shift registers; encapsulated logic elements; custom encapsulation service; preset electronic counters; transistorized indicator lights*.

C W S Waveguide Corp., Booths 1313-1315

301 West Hoffman Ave. Lindenhurst, L.I., N.Y.

Carl W. Schutter, Ing Bian Oei, V. Schutter, S. Amir, J. Ashman

Dunniny Antenna—Coasial Connectors & Switches—Couplings—Directional Couplers— Crystal Mounts—Duplexers—Adapters—Rotary Joints—Horns—Rigid Waveguides,





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World Radio History

Burnell & Co. Inc. 10 Pelham Pkwy. Pelham Manor, N.Y. Booths 2909-2910 ▲ L. G. Burnell, ▲ N. Burnell, L. Schwartz, Julius Tischkewictsch, Ray Bello, Marty Nemi-roff, ▲ Bernie Teinerman, Nat Cohen



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Burr-Brown Research Corp., Booth 3052 Box 6444

Tucson, Ariz. ▲ Thomas R. Brown

Transistorized Instrumentation for laboratory and field use-Operational amplifiers, Instru-mentation amplifiers, Oscillators, Voltmeters, Packaged for plug-in rack mounting or as in-dividual instruments.

(Continued from page 205A)

Burroughs Corporation Electronic Tube Div. P.O. Box 1226 Plainfield, N.J. Booths 1211-1215 A. Shesser, ▲ S. Kuchinsky, ▲ R. Wolfe, ▲ J. Bethke, R. Brady, P. Nace, R. Driscoll, G. Arndt



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Burton-Rogers Co., Booth M-17 See: Hoyt Electrical Instrument Works,

▲ Indicates IRE member Indicates new product.

Bussmann Mfg. Division McGraw-Edison Co. University at Jefferson St. Louis 7. Mo. Booth 2737 A. L. Branning, C. J. Dane, T. P. Lawless, A. H. Lucas, J. D. Rambo, E. F. H. Harvell, F. M. Sibley



BUSS Add-On Fuse Blocks

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(Continued on page 20821)

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Burlingame Associates 510 South Fulton Ave. Mount Vernon, N.Y. Booths 3811-3816 Boollis 5811-5810 A Harold A. Bogin, A Roland Reisley, A Gerry Sullivan, A Arnold Ackerman, A Don Anderson, A Al Beckman, A Don Carter, A Bob Crane, A Bernie Green-berg, A Harold Kurland, A Al Lee, A Marty Maloy, A Jim Randall, A Joe Giguere, A Clark Smoll, A John Takak-jian, A Dick Bullock, A Jim Bogin Coordinated Test Equipment Display-Coordinated rest Equipment Display— Also New instruments for: precise gen-eration of AC and DC Power, Voltage; Accurate AC and DC Voltage measure-ment and calibration.

Burndy Corp., Omaton Div. Norwalk, Conn. Booths 1329-1331

S. Bergman, A. Aune, L. Gage, E. Val-enrach, R. Smith, A. Behnke, S. Schul-man, W. Bonwitt, H. Dupre, M. Lazar, D. Dibner, F. March, L. Gray, L. Berk-ley, S. Cotro, P. Putignano, M. Potenza, R. Atkinson, E. Garwett, R. Resker, E. Salz, G. Woeth, P. Costello, J. Cos-tello, P. Carwithen, M. Elkind, J. Ber-tram, M. Gregory, G. Turrian Complete Line Solderless HYFEN conmond, W. Gregory, G. Turran Complete Line Solderless HYFEN con-nectors, crimp-type snap-locked contacts: Edge Type, puntype printed circuit HYFENS, *HYFEN receptacle for plug-in components; coax HYFENS for recu-lar corx, miniature coax; rack, panel HYFENS. Uniring, Modulon, Stapin lines, Related hand, semi-automatic, & au-tomatic tooling. tomatic tooling

(Continued on page 206.1)

▲ Indicates TRE member.

* Indicates new product.

- First and Second floors-Components
- Third floor—Instruments and Complete Equipment
- Fourth floor—Materials, Services, Machinery

For basic research, pilot plant studies and the more efficient production of semi-conductors-new Lindberg Diffusion Furnaces



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(Continued from page 202A)

British Radio Electronics Ltd., Booth 1820

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Precision Linear & Sine/Cosme Potentiometers; Silvered Mica Capacitors; Miniature Air Dielectric Temperature Compensating Capacitors; Sub Miniature Deposited Carbon Resistors; Switches and Transformers; Slow Motion Tuning Dials.

Brubaker Electronics, Booth 2128 See: Telecomputing Corp.

Brush Instruments Div., Booths 2616-2626

See; Clevite Corp.

Buchanan Electrical Prods. Corp. 225 U.S. Route 22 Hillside, N.J. Booth 2341 Johnson, Albert C. Schultz.

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A. W. Amundson Electroformed mesh and evaporation masks for electron tubes and subminiature transistor manufacturing, etched metal and glass precision parts for special projects or mass produced component parts, precision printed circuits, micro mesh sieves for submicron particle size determination.

Budd Company, Booths 4224-4226

See: Continental-Diamond Fibre Corp.

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Indicates new product.

Hewlett-Packard Electronic Sweep Oscillators are new measuring tools deliberately designed to give you simpler, faster microwave measurements. Four models are provided, covering frequencies 2.0 to 18.0 KMC as follows: Model 683A, 2.0 to 4.0 KMC; Model 684B, 4.0 to 8.1 KMC; Model 686A, 8.2 to 12.4 KMC and Model 687A, 12.4 to 18.0 KMC.

These instruments make possible microwave investigations and evaluations with a convenience previously associated only with lower frequency measurements. These oscillators provide a wide range of sweep speeds so that measurements of reflection, attenuation, gain etc., can be displayed on an oscilloscope or recorded in permanent form on X-Y or strip-chart recorders.

Electronic Sweeping

Specifically, the new oscillators provide either a CW or swept rf output throughout their individual bands. The instruments employ new backward wave oscillator tubes whose frequency is shifted by varying an applied potential. Thus, troublesome mechanical stops and tuning plungers are eliminated. Sweep range is continuously adjustable and independently variable; sweep rate is selected separately, and either can be changed without interrupting operation. The full band width can be covered in time segments ranging from 140 seconds (very slow for mechanical recorder operation) to 0.014 seconds (high speed for clear, non-flickering oscilloscope presentation).

Linear Frequency Change

The swept rf output from the \oplus sweep oscillator is linear with time, and a linear sawtooth voltage is provided concurrent with each rf sweep to supply a linear time base for an oscilloscope or recorder. In addition, for convenience in recording and other operations, rf sweeps can be triggered electrically externally and single sweeps can be triggered by a front panel push button. The rf output can also be internally AM'd from 400 to 1,200 cps and externally AM'd or FM'd over a wide range of frequencies.

Rapid Visual Presentation

The variety of sweep rates and band widths available from the new oscillators insures convenience and accuracy for reflection and transmission coefficient measurements and many other production line and laboratory tests. For maximum speed, an oscilloscope such as \oplus 130A/B may be used as indicated in the diagram on opposite page. For maximum information and a permanent record, an X-Y or strip chart recorder may be used.

Complete details of a rapid visual method using an oscilloscope or a maximum-data, permanent record method using a recorder may be obtained from your @ field engineer. Detailed discussions of these methods are also contained in the @ Journal, Vol. 8, No. 6, and Vol. 9, No. 1-2, available on request.

TYPICAL SPECIFICATIONS

Below are specifications for -hp- 686A Sweep Oscillator, 8.2 to 12.4 KMC. Specifications for -hp- 683A, 684B, and 687A (P band) are similar except for frequency range and other minor variations.

Types of Outputs: Swept Frequency, CW, FM, AM.

Single Frequency Operation

Frequency: Cantinuausly adjustable 8.2 to 12.4 KMC.

Power Output: At least 10 milliwatts inta matched waveguide load. Cantinuausly adjustable ta zera.

Swept Frequency Operation

Sweep: Recurrent; externally triggered; also manually triggered single sweep. Rf sweep linear with time.

Power Output: At least 10 MW inta matched waveguide laad. Output variatian less than 3 db aver any 250 MC range; less than 6 db aver entire 8.2-12.4 KMC range.

Sweep Range: Adjustable in 7 steps 4.4 MC ta 4.4 KMC.

Sweep Rate-of-Change: Decade steps fram 32 MC/sec. ta 320 KMC/sec.

Sweep Time: Determined by sweep range and rate; fram 0.014 to 140 seconds aver full-band.

Sweep Output: +20 ta +30-valt-peak sawtaath provided at a frant-panel cannectar cancurrent with each rf sweep.

Modulation

Internal Amplitude: Square wave madulatian cantinuausly adjustable fram 400 ta 1200 cps; peak rf autput pawer equals cw level \pm 1 db.

External Amplitude: Direct caupled to 300 KC; 20 valt swing reduces rf autput level fram rated cw autput to zero.

External Pulse: +10 valts or mare, 5 millisecond maximum duratian.

External FM: Apprax. 350 v peak ta modulate full frequency range.

General

Input Connectors, Impedances: BNC; abave 100,000 ahms.

Output Connector: Waveguide cover flange (686A, 687A); Type N, female (683A, 684B).

Power Requirements: 115/230 volts $\pm 10\%$, 50/60 cps; appraximately 540 watts.

Price: (*) 683A (2.0 ta 4.0 KMC) \$3,000.00. (*) 684B (4.0 to 8.1 KMC) \$2,900.00. (*) 686A (8.20 ta 12.40 KMC) \$2,900.00. (*) 687A (12.40 ta 18.00 KMC) \$3,400.00.

> (Prices above are f.o.b. factary for cabinet models. Rack maunt instruments \$15.00 less.)

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Figure 1. Arrangement for high speed microwave meas-

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200A

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Whom and What to See at the Radio **Engineering Show**

(Continuea from page 196.4)

Boonton Electronics Corp. 738 Speedwell Ave. Morris Plains, N.J. Booth 3114 ▲ Ernest A. Porter, ▲ John H. Mennie, J. Frucht, ▲ John M. Young, ▲ Earle Dilatush, ▲ Charles Seldon, ▲ Vincent A. Schauler



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Boston Division, Booth 2208 See: Minneapolis-Honeywell,

Boston Insulated Wire & Cable Co., Booth 4017 65 Bay St.

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T. W. Wise, W. S. Aldrich, R. D. Adams, F. C. Kluhsman, J. J. Duffy, W. Lawrence, R. W. Trautlein, C. M. Byron, M. Kennedy, G. Famy Iratitein, C. M. Byron, M. Kennedy, C. Famy Wire Markers, Special Makers, Naneplates, Narrow Slit Tapes, Mini-Tapes, for marking, holding, bundling, instructing in the manufacture of electrical and electronic equipment; Markermatic automatic marking machine; Printmatic label printer-diecutter; Tapematic Pre-cut tape dis-penser; Printed circuit tapes and shapes.

(Continued on page 202A)



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VISIT US AT IRE BOOTH #1218-1224 March, 1960

World Radio History



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Bomac Laboratories, Inc. **Div. of Varian Associates** Salem Rd. Beverly, Mass. Booths 2710-2712 Cyrys Haller, James F. Dean, A Robert W. Andrews, A Bayard Robbs, Williard S. Ferris, C. Peter Roberts, Earl D. Benson



Microwave Tubes and Components, (Continued on page 198A)

▲ Indicates IRE member. * Indicates new product.

See all the exhibits!

Don't miss these important locations-

Mezzanine at back of first floor, South Room at center of south wall, second floor, 3000 court at southeast corner of third floor, 4000 court at southeast corner of fourth floor. 4500 court in northwest corner of fourth floor.



62 PEARL STREET, ATTLEBORO, MASS.

(Continued from page 190.4)

James G. Biddle Co., Booth 3943 1316 Arch St. Philadelphia 7, Pa. Paul P. Emery, K. Fugit, E. E. Lange, A. Q. Lange, O. X. Heinrich, W. G. Long, P. E. Sellers, S. C. Sommer

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(Continued on page 196A)

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This instrument's many outputs and different modes of operation, coupled with excellent frequency stability and high output (10v) over the entire frequency range, make it the most versatile audio-video test instrument commercially available.



See a Typical Standards and Measurements Laboratory in Operation . . . Impedance Measurements from D-C to Microwave Frequencies

The New Beat-Frequency Generator Will Be on Display as well as many other instruments

Type 1300-A Beat-Frequency Video Generator ... \$1950.

As Monuolly-Tuned Generator: Sine Wave: 20c to 12 Mc Square Wave: 20c to 2 Mc As Sweep Generator (60c sweep rate): Sine Wave: 20 kc to 12 Mc Sweep width is continuously adjustable from 0 to ±6 Mc at any center frequency from 0 to 12 Mc. Horizontal deflection voltage and blanking pulse provided for scopes. Colibration Accuracy: 20c to 20 kc, ±(1% + 1c) 20 kc to 500 kc, ±(1% + 1 kc)

In-ade	lition to the main frequency di	ial,
two	increment dials calibrated fro	оm
-5	Do to ± 50 c, and -20 kc to ± 20	kc.
are	provided. Calibration accurac	ies
are	±1c and ±0.5 kc, respectivel	٧.

Sine Wave — harmonic dislottion 20c to 20 kc: < 1.5% of output 20 kc to 12 Mc: < 4% of output

Square Wave Rise time less than 0.075 μsec above 300 kc Top flat to 2% of peak-to-peak at 60c, 5% at 20c. Hum: less than 0.1% of output

	Voltag Sine-Wave (rms)	e Range Square-Wave (peak-to-peak)	Accuracy	Frequency Characteristic	Output Impedance
Attenuator output	0.1, 0.3, 1, 3, 10, and 30 mv; 0.1, 0.3, and 1v full scale,	1, 3, 10, 30, 100, and 300 mv; 1, 3, and 10v open circuit	= 3% of full scale; attenuator db incre- ments = 1%	flat within ± 0.25 db from 40c to 20 kc (± 0.75 db at 20c); ± 1 db from 20 kc to 12 Mc	75Ω ≠ 2 9
High output	0 to 10v	0 to 10v	±3% of full acale	flat within ± 0.25 db from 20c to 20 kc; ± 1 db from 20 kc to 12 Mc (open circuit)	820Ω ± 25

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