

JANUARY, 1967 75¢



the magazine of electronic servicing

COLOR

- Removing the Barber Pole
- Symbol Standardization
- Ohm's Law for AC
- Industrial Electronics Notebook
- Symfact

Going to profit here?

Then profit here.



82 Channel TAC-4 List Price \$72.50

VHF Only Model TC-88 List Price \$39.95

New Jerrold 82 Channel Coloraxial Home System makes perfect sales mate with Color TV Antennas

Look around you. Every homeevery new housing developmentis a potential profit-making sales spot for you and the TAC-4 Home Master TV System. Here's why:

TAC-4 makes good color TV go a long way. Every color TV set needs a top-grade antenna for good reception. TAC-4 makes this good reception possible anywhere in the house by providing extension antenna outputs in various rooms for portable TV and FM sets as well as extra sets. You can tie-in a TAC-4 with every antenna sale. And don't forget, it's a handy sales feature that housing developers like to use to help sell a home.

TAC-4 handles all 82 TV channels (color and black and white) as well as FM. It's the industry's first amplified coupler that can handle all present and future channels. Just connect the antenna to the input and then connect four or more sets to the outputs.

TAC-4 is easy to install. Coaxial inputs and outputs make connections, simple. Coaxial cable can be run right along with other electric wires—without interference with the signal. And call backs are practically unheard of because the amplifer is completely solid-state.

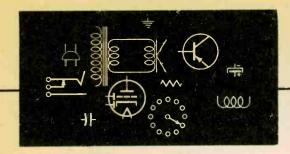
TAC-4 pleases customers. It's the first truly professional installation for assuring better TV and FM reception anywhere in the house. It's the superior amplifier-coupler to sell with any TV set or antenna such as the Jerrold Pathfinder series. Sell them together. Customers get top reception. You get top profits.

For more details, see your distributor.



401 Walnut St., Phila., Pa. 19105

Circle I on literature card



Symbol Standardization

As a publisher in the electronics industry, Howard W. Sams & Co., Inc., has a vital interest in the standardization of schematic symbols. There is at present an industry-wide program to accomplish this. For many years now, each manufacturing company and sometimes, we suspect, each individual draftsman, had a different concept as to the proper method of symbolization.

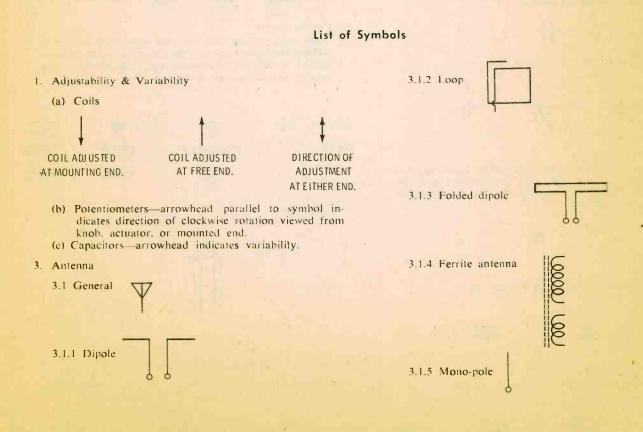
The Electronic Industries Association (formed by participating electronics manufacturing concerns), in cooperation with the American Standards Association, has now published a list of graphical electronic symbols with the objective of simplification of symbols, reduction of drafting time, and a better understanding of schematics by all technicians.

We are reproducing a portion of the list as an aid to the technician. We feel that if all manufacturers and publishers were to adopt these symbols as standard, the time spent by technicians in interpreting symbols could be reduced greatly.

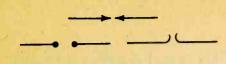
The Howard W. Sams & Co., Inc., is adopting these standard symbols in all of its publications.

Definitions and Interpretation of Symbols

- Symbol—A symbol is the aggregate of all its parts, but to improve readability, parts of a symbol, such as a multisection capacitor, may be separated.
- Orientation of Symbol—The orientation of a symbol one drawing, including a mirror image, does not alter the meaning of a symbol.
- Arrow Heads—Unless otherwise noted, no significance is placed on open or closed arrowheads.
- Angles of Lines—Generally, the angle at which a connecting line is brought to a graphical symbol has no particular significance unless otherwise noted.
- Width of Lines—The width of the line does not affect the meaning of a symbol.



PF REPORTER, January, 1967 Vol. 17, No. 1 PF REPORTER is published monthly by Howard W. Sams & Co., Inc., 4300 W. 62nd, Indianajolis, Indiana 46206, Second-class postage paid at Indianajolis, Indiana, 1, 2 & 3 year subscription prices; U.S.A., its possessions and Canada; \$5.00, \$8.00, \$10.00, Other countries: \$6.00, \$10.00, \$13.00, Current single issues 75¢ each; back issues \$1, each.



7. Battery

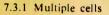
The long line is always positive, but polarity may be indicated in addition.

7.2 One cell

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7.3 Two cells

키바



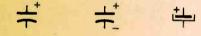


8. Capacitor

8.1 General—If it is necessary to identify the capacitor electrodes, the curved element shall represent the outside electrode in fixed paper-, mica-, ceramic-, and plastic-dielectric capacitors; the moving element in adjustable and variable capacitors; and the low-potential element in feedthrough capacitors.



8.1.1 Polarized capacitor (electrolytic)



8.1.2 When multiple-section, electrolytic capacitors are diagrammed, each section must be identified with the appropriate symbol to correspond with the identifier used on the actual part.



8.1.2.1 Nonpolarized electrolytic capacitor.



8.1.2.2 Multiple sections in common container.





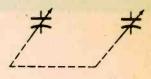




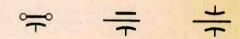
8.1.3 Adjustable or variable capacitor.

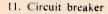
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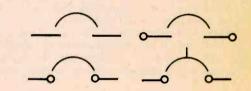
8.1.4 Adjustable or variable capacitors with mechanical linkage of units.



8.1.5 Feedthrough capacitor.







13. Circuit returns

The rake symbol may be used to indicate accessible metal ground, earth ground, the chassis, or where chassis ground is different or isolated. The other symbols may indicate ground, chassis ground, or B—.



16. Coil, relay



18. Connector-disconnecting device.

18.1 Female contact, commonly used for a jack or receptacle (usually stationary). 18.2 Male contact, commonly used for a plug (usually movable).

18.8.2 Male contact.

• OR 🚥

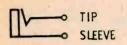
18.3 Separable connectors (shown engaged).

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18.6 Communication (telephone) type connector.

4

18.6.1 Two-conductor jack.



18.8.3 Two-conductor nonpolarized connector with female contacts.

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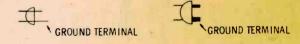
18.8.4 Two-conductor nonpolarized connector with male contacts.



18.8.5 Two-connector polarized connector with female contacts.



18.8.6 Two-conductor polarized connector with male contacts.



18.9.1 Phono-type connector plug.

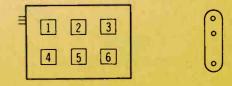


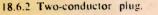


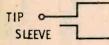
18.9.2 Phono-type jack

2

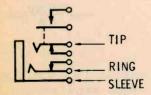
18.9.3 Wire end of female socket (cap)





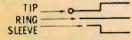


18.6.3 Three-conductor jack with two break contacts (normals) and one auxiliary make contact.



18.6.4 Three-conductor plug.

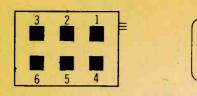
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18.8 Connectors of the type commonly used for powersupply purposes (convenience outlets and mating connectors).

18.8.1 Female contact.

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25. Core

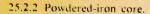
25.1 General or air core

If it is necessary to identify an air core, a note should appear adjacent to the symbol of the inductor or transformer.

25.1.2 Adjustable core.



25.2.1 Laminated core.



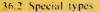
25.3 Core of electromagnet.





36. Fuse.

36.1 General.



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42. Inductor.

42.1 General

Use symbols as shown unless otherwise indicated or required by special considerations.



42.2 If it is desired especially to distinguish magneticcore inductors:

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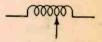
42.3 Tapped, air core.



42.4 Adjustable inductor: arrowhead points in direction of adjustment location. Bottom of coil is at mounted end.

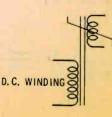


42.4.1 Adjustable inductor, air-core: arrowhead indicates movable tap. Unless noted, coil mounting end not specified.



42.7 Saturable-core inductor (reactor).

Polarity marks may be added to direct-current winding. Explanatory words and arrow are not part of the symbol shown.



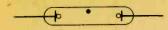
44. Lamp.

44,1 Ballast lamp (tube)

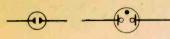
The primary characteristic of the element within the circle is designed to vary nonlinearily with the temperature of the element.



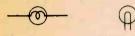
44.2 Lamp, fluorescent.



44.3 Lamp: glow, cold-cathode, neon.



44.4 Lamp, incandescent.



- 46. Machine, rotating.
 - 46.2 Generator.

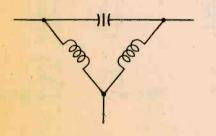
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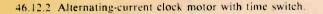
46.3 Motor.

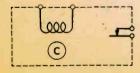
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46.12.1 Alternating current reversible motor.







46.12.3 Phono motor.



48. Meter-instrument.

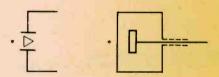
Note—The asterisk is not a part of the symbol. Always replace the asterisk by one of the following letter combinations, depending on the function of the meter or instrument, unless some other identification is providedin the circle and explained on the diagram.

DB or DBM DB (decibel) meter MA Milliameter O R Ohmmeter Recording meter T V Tuning meter Voltmeter VOM Voltohmmeter Vacuum tube voltmeter VTVM Volume indicator, audio level meter VU W Wattmeter



51. Microphone.

51.1 Ultrasonic microphone or transducer. * Iudicate type by note: ceramic, crystal, dynamic, etc.



\$1.2 Audio-frequency microphone. * Indicate type by note.



56. Oscillator, signal generator, or unspecified alternatingctirrent source.



58. Witting.

58.5 Crossing of paths or conductors not connected. The crossing is not necessarily at a 90° angle.



58.6.2 Junction of connected paths, conductors, or wires.





58.8.3 Two-conductor cable.

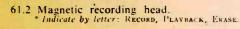


58.8.4 Shielded two-conductor cable, shield grounded.

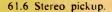


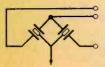












62. Piezoelectric unit.

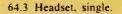


64. Receiver-earphone. headphone.

64.2 Headset, double.







65. Rectifier, semiconductor.

65.1 General.

Note: Triangle points in direction of forward (easy) current as indicated by a DC ammeter, unless otherwise noted adjacent to the symbol. Electron flow is in the opposite direction.

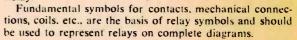
65.2 Semiconductor rectifier.



65.2.2 Silicon controlled rectifier (SCR).



66. Relay.



66.2 Relay coil:





66.3 Relay contacts.



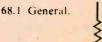


TRANSFER

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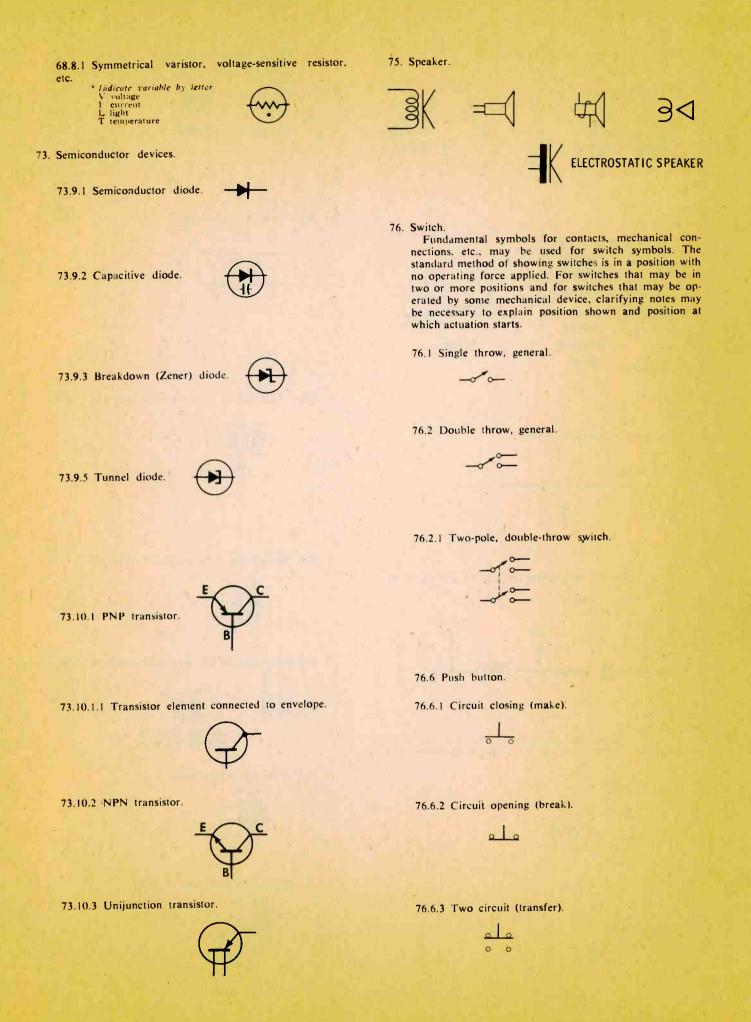
68. Resistor.

BREAK

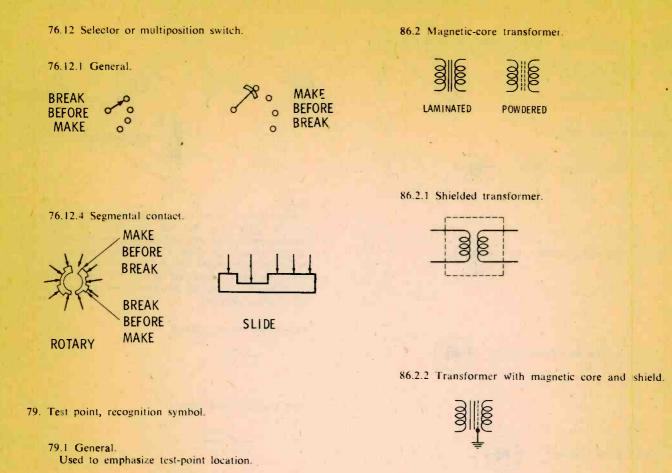


68.2 Tapped resistor.

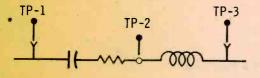
68.3 With adjustable contact.



10



79.2.1 Test-point recognition for test jack and circuit terminal. * Reference designation. Not part of symbol.



84. Thermistor.

"T" indicates that the primary characteristic of the element within the circuit is a function of temperature.

84.1 General.



86.1 General

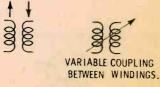


86.3 One winding with adjustable inductance

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86:4 Each winding with separately adjustable inductance.



86.6 With taps, single phase.



86.7 Autotransformer, single phase.



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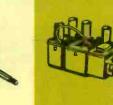
U-V







COLOR





TRANSISTOR

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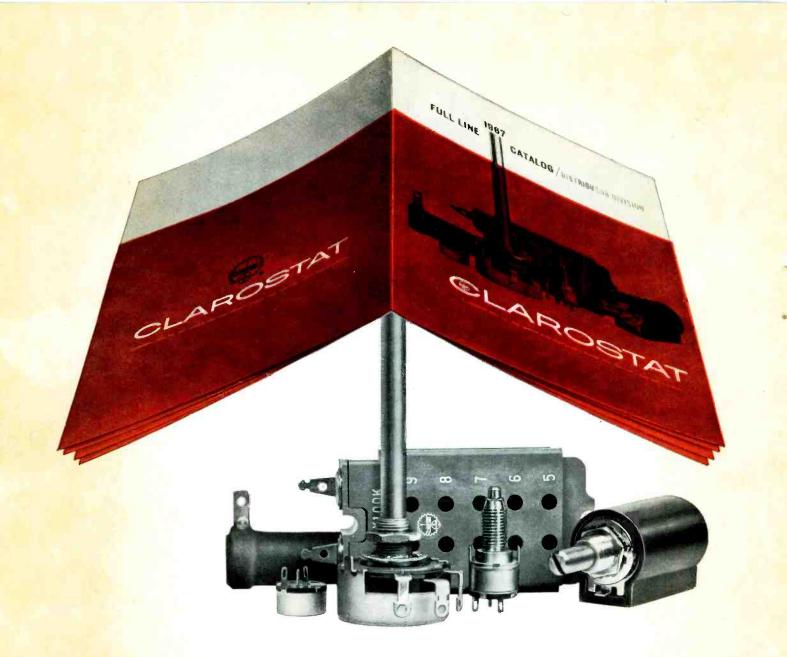
GUARANTEED COLOR ALIGNMENT - NO ADDITIONAL CHARGE

Simply send us the defective tuner complete; include tubes, shield cover and any damaged parts with model number and complaint. Your tuner will be expertly overhauled and returned promptly, performance restored, aligned to original standards and warranted for 90 days.

UV combination tuner must be single chassis type; dismantle tandem UHF and VHF tuners and send in the defective unit only.

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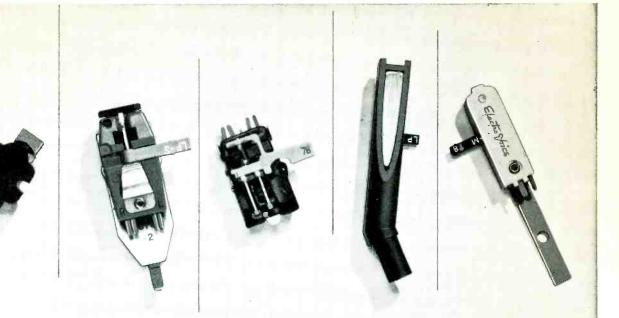


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Indexed in Lactrodax. Printed by the Waldemar Press Div. of Howard W. Sams & Co., Inc.

edo PHOTOFACT

the magazine of electronic servicing

VOLUME 17, No. 1

JANUARY, 1967

CONTENTS

Symbol Standardization A list of standard electronic symbols approved by the Electronics Industries Association		1		
The Electronic Scanner		1.5		
Intermittent and Time Lapse Sync Troubles Case histories reveal causes and effective troubleshooting procedures.	Allan F. Kinckiner	17		
Removing the Barber Pole An understanding of circuit functions and symptoms, together with a systematic isolation procedure, quickly solves color sync failures.	Robert F. Heaton	22		
The Case of the Bifilar Balun Rewinding solves the problem of no replacement.	Jack Darr	24		
Depreciation Reserves For Equipment An introduction to the various methods of computing depreciation on test equipment, service vehicles, and other forms of tangible assets.	Ralph H. Butz	26		
Ohm's Law For AC A review of AC circuits involving Ohm's law, Pythagorean formula, power factor, and component characteristics.	Robert G. Middl <mark>e</mark> ton	28		
Book Review		37		
Notes on Test Equipment Lab reports on Mercury Model 2000 Tube Tester and RCA Model WR-508 RF Signal Generator.	T. T. Jones	38		
Symfact: Multiplex Amplifier and Doubler Operation and trouble analysis.		41		
Industrial Electronics Notebook—Part I Becoming familiar with electronic circuit applications in industry.	Ed Bukstein	45		
The Troubleshooter		57		
Product Report		61		
Color Countermeasures		68		
Free Catalog and Literature Service		72		
Monthly Index on Free Literature Card				

About the Cover

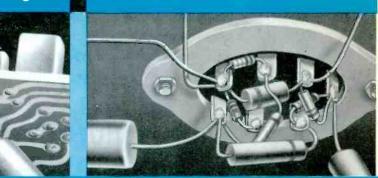
The technician with the frown is evidently having a bit of trouble "removing the barber pole". He won't get rid of it with that channel selector knob. We'll bet, however, that he'll never again have difficulties with chroma sync after reading the article starting on page 22.

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...lift a wire-lead component from a printed wiring board for testing

... test or replace a capacitor or resistor on a crowded tube-socket



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WORLD'S LARGEST MANUFACTURER OF CAPACITORS Circle 5 on literature card

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The Electronic Scanner

news of the servicing industry

Potpourri

The Electronics Industries Association's Citizens Band Section recently endorsed the efforts of the Federal Communications Commission in combating violations in Citizens Band Radio operations, and pledged full cooperation in aiding the FCC to solve the "growing problem of enforcement in the administration of Citizens Band operations."

It pledged an "affirmative approach" to the potential user in advertising and through distribution of "dos and don'ts" literature to purchasers of CB equipment, and encouragement of salesmen to provide guidance in the proper use of the equipment.

A nationwide program to support servicing retailers has now been set into operation by **Motorola Con**sumer **Products.** A "Technical Training Rep." force of 50 men is now strategically located across the nation to train Motorola dealers and service companies so that they, in turn, will be better equipped to provide customer service. The territorial assignments are based on population density with an additional training task force of four men operating out of Chicago.

The "Technical Training Representatives" offer personalized, upgrading training sessions for consumer electronic service technicians in their place of business, with scheduling arranged by regional service managers. These individualized training sessions are scheduled for one full day or more.

S. R. "Ted" Herkes, president of Motorola Consumer Products, observed that the industry is "on the eve of the greatest technological advancement since the advent of television and the world ahead will be a solid-state world." Consumer electronic technicians must be trained to be ready to handle the products of this new era, he added.

Once trained in solid-state products, perhaps servicing dealers will rightfully gain some deserved recognition for their knowledge and investment and the essential role they perform, Herkes added.

Muntz TV has announced plans to change its name on January 1, 1967 to Television Manufacturers of America Co. The company explained to stockholders that the changing nature of the firm's business now calls for a more fitting name and identity. A substantial share of the company's sales now results from production of television and stereo brands other than the Muntz brand, including private label television sets. However, sale of Muntz brand television sets and stereo units will continue.

www.americanradiohistory.com



"My shop's been loaded... since I got my FCC License."

"And I could kick myself for not getting it sooner. I'm pulling in all kinds of mobile, marine and CB business that I couldn't touch before; have even had some calls to work on closed-circuit television. I've hired two new men to help out and even with them, I'm two weeks behind."

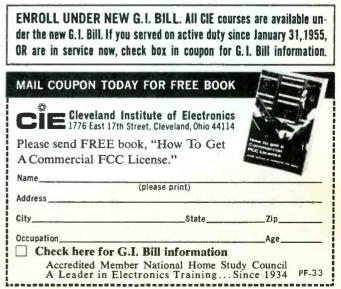
And so it goes. Once you have that FCC ticket, you open the door to all kinds of new business. And that's not all. The knowledge you need to pass the FCC exam gives you a fundamental understanding of *all* electronics. You'll find you can do more work in less time...work on almost *any* kind of electronics gear.

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Sales of FM radios will soar to new highs in 1967, according to a specialist in the radio receiver field. William B. Keepin, manager of the **Norelco** radio department of North American Philips, predicts sales of close to 12 million FM sets during the coming year—a 20% gain over 1966 estimated sales of nearly 10 million units. Currently this country's radio-listening public owns about 34 million FM receivers.

Expansions

A three-year \$20 million program to expand the development and manufacture of semiconductor products was announced by **ITT Semiconductors Worldwide.** The \$20 million expenditure will be divided approximately 50% in the United States and 50% overseas. Included in the program will be the updating of semiconductor facilities in Lawrence, Mass. and Palo Alto, Calif., and a \$5.3 million expansion of the present West Palm Beach semiconductor manufacturing plant.

Electrohome of Canada recently opened its new \$4,-000,000 cabinet plant in Kitchener, Ontario.

The entire 450,000 square foot plant houses computerized and automated production systems for precision cabinet design. Electrohome's 500 craftsmen now work with electronic woodworking devices that enable one worker to perform ten operations within minutes. With the opening of the new plant, skilled technicians and modern machines can now deliver a completed TV, Stereo, or Hi-Fi cabinet every 30 seconds.

An appropriation of approximately \$2 million for an immediate expansion of compactron facilities has been approved by the board of directors of **General Electric.** According to I. D. Daniels, Manager of Manufacturing for the Tube Department, "Compactrons should start coming off the first additional compactron production lines by January 1967, and all lines are scheduled to be in full production during the second half of the year."

Sales

Record third quarter and nine month sales and a sharp increase in earnings were reported by **Admiral Corporation.** Consolidated sales in the third quarter were \$110,425,383, 58% higher than 1965 sales of \$69,826,396, and were also a record for any quarter. Earnings after taxes increased 95% to \$2,270,356, up from \$1,163,378 last year.

President Vincent Barreca said that substantial expenses in connection with major moves of military and consumer product manufacturing facilities and their related start-up costs were charged off during the third quarter, and the earnings for the 9-month period were 139% above last year.

• Please turn to page 70



Circle 8 on literature card



Intermittent and Time Lapse Sync Troubles

Intermittent sync troubles, like most intermittent troubles, are often due to loose connections, unwanted contacts between wiring or components, or mechanically damaged components. It is perfectly normal for these conditions to occur.

In some receivers made between 1950 and 1955 terminal strips are fastened by soldering the mounting legs of the strips to lances punched in the chassis. As these sets became older, soldering operations on the wiring end of the mounting-leg terminal affected the soldering mounting and sometimes introduced cold joints. Another manufacturer used hollow sleeves riveted to the chassis for ground connections. The hollow sleeves were actually elongated hollow rivets; however, they were fragile, and the pressure necessary for metal-to-metal bonding resulted in slightly resistive connections. Vir-

(A) V.O.-Vertical-oscillator pulse

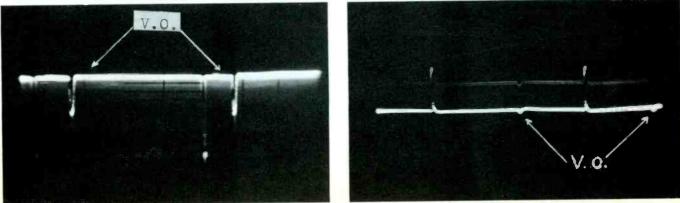
tually every set maker has used tubesocket mounting rivets that served as ground contacts; the fiber shrank with age, the rivet became loose, and, again, a resistive connection resulted.

by Allen Kinckner

These conditions affect many phases of receiver operation and have continually plagued experienced servicemen. The most exasperating cases are those where picture stability is affected by a resistive connection that introduces distortion to the sync signal. One of the doggiest cases I can remember involved a Westinghouse Chassis V-2233-4. It had intermittent horizontal quivers, sometimes slight, sometimes more severe, and sometimes a complete tearout. Signals fed to the AFC tube, both the sync and the sample, looked normal when viewed at one-half horizontal frequncy, but when the same signals were scoped at one-half vertical frequency, a 60-cps hum was readily noted. The trouble was pinpointed to a resistive solder connection on the lance that mouted a terminal strip and also served as a ground for various components in the AFC circuit. Two mysteries surrounded this case: The bad connection also served as a ground for the verticaloutput tube filament, but vertical deflection was not affected. The hum at varying levels was present at all times, but the trouble was intermittent. This last point, steady signal distortion but intermittent effect, is one of the imponderables encountered in servicing intermittent sync troubles.

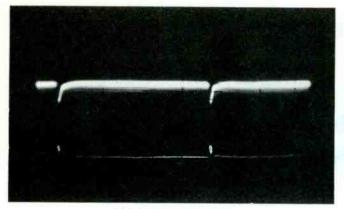
Localizing Source of Trouble

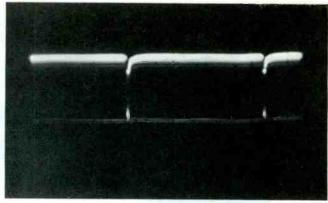
The general practice in scoping is on a stage-to-stage basis. In scoping sync systems, especially those with more than one stage, it is best to first scope the input signal and



(B) Feedback pulse negative

Fig. 1. Oscillator pulse feeds back to sync output if integrator output is okay.





(A) Perfect sync signal
 (B) Average sync pulse
 Fig. 2. Perfect sync signal has even tip level, average signal is uneven.

then move to the output of the final stage. This procedure has several advantages. If trouble is originating within the sync system, it will show up as adequate or distorted sync signals. A normal sync signal indicates trouble in either the deflection oscillator, sync coupling capacitor, or vertical integrator. Condition of the integrator can be checked at the sync output. For example, if the sync pulse is getting through the integrator to the vertical oscillator, then the vertical oscillator signal will pass through the integrator and appear in the sync output as shown in Figs. 1A and 1B. These traces are best obtained by using line rather than internal sync for the scope.

A perfect sync signal, Fig. 2A, can only be obtained in receivers having some method of clipping the sync tips. The average sync signal is likely to be fuzzy at the tip level (Fig. 2B), indicating noise or variations in the tips of the horizontal sync. Both conditions are normal and are not representative of a fault that will affect picture synchronization. However, cases of poor synchronization have been found when only a small amount of distortion was noticeable in the sync signal.

An RCA KCS68 had intermittent vertical sync and jitters. Sporadically the vertical-blanking bar would wander about half way down the raster (Fig. 3) then snap out of sight. Sometimes the picture would jitter slightly, or often it would lock perfectly. Scoping at the input grid of the vertical separator, pin 4 of V13 in Fig. 4, gave the normal trace in Fig. 5A. Moving the scope to the sync output, pin 5 of V15, showed the sync pulse was missing. Backtracking to the plate, pin 5 of V13, also revealed no sync signal. These conditions were noted when the vertical-blanking bar was wandering. During jitters, small sync signals were noted. When the picture was locked normally, W5B and W5C were found on the plates of V13 and V14, respectively. These waveforms indicated trouble in V13. but voltages on this tube didn't vary as the different troubles occurred. Probing around V13 revealed that the trouble was due to a loose connection of C75.

As was previously mentioned, a scope check of the sync output will indicate if signals are getting through the vertical integrator. You can check to see if sync is reaching the horizontal oscillator by scoping at the output side of the coupling capacitor. Sets using a synchoguide oscillator will display the signal shown in Fig. 6A. One case of infrequent horizontal tearout would not recur in the shop, even though the set was rack-checked for almost two days. Scoping the output side of the sync-coupling capacitor (which in synchoguides is the grid of the AFC stage) showed the crazy trace in Fig. 6B. The problem was caused by a defective filter capacitor. The interesting point is that the picture remained in sync even though the sync signal was greatly abnormal.

Tube Tester Doesn't Always Tell

Even when intermittent sync trouble is caused by a tube, a cure is rarely possible on a service call. For example, if the trouble occurs at two- or three-minute intervals,

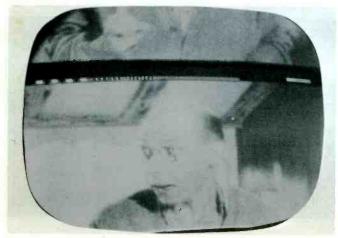
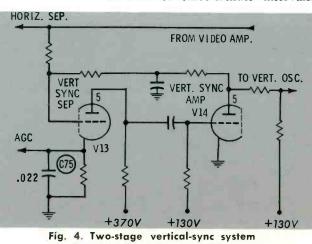
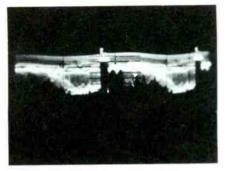


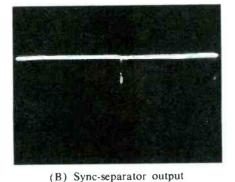
Fig. 3. Moving vertical-blanking bar.

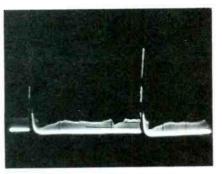


found in RCA Chassis KCS68.



(A) Sync-separator input





(C) Sync-amplifier output

Fig. 5. Appearances of the vertical-sync pulses taken at different points in the sync-separator circuit.

replacing the suspected tube(s) may result in normal operation only for the limited time allowed for house calls. In many instances, it will be found that the new tube(s) only increases the time between trouble appearances.

Intermittent rolling for short periods (about once or twice an hour) can really be a "dog" to locate. A fellow serviceman asked me to look at a Crosley 487 with this disease. I knew he was a thorough troubleshooter, one who substitutes all tubes, so I expected rough sailing. Once again the scope proved to be a time-saver. The sync-separator plate (the triode section of 6BA8) revealed the trace in Fig. 7A, but the grid of this tube showed a normal signal. The normal signal in Fig. 7B was obtained when the 6BA8 was substituted.

The defect in this 6BA8 is not revealed by a tube tester. The name of the trouble is *radiation*. The signal in the video-amplifier pentode

section radiates or couples into the triode section.

Radiation

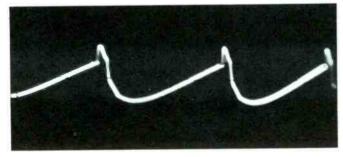
About two weeks after the incident with the Crosley, the same technician invited me to look at another set with the same infrequent rolling. My friend had already tried new tubes, and scope traces throughout the set were normal. With the LC probe on the separator plate, the normal signal distorted when I moved the set. As shown in Fig. 8, the vertical-sync pulses virtually disappeared. I made sure the probe contact was good, and then found that by flexing the chassis the trouble could be made to appear. The defect was eventually traced to a loose mounting screw holding the voke.

An electrical analysis of the trouble revealed that with poor contact between the mounting strap and the yoke core a signal developed. This signal was being picked up by the sync-separator grid. The highimpedance grid circuits of sync stages are prone to picking up radiation of unwanted signals originating in close proximity.

An RCA using a 6BU8 also presented a case of yoke radiation that caused intermittent vertical rolling. My first thought was that the recently replaced yoke was the cause. Then I noticed the replacement yoke had excessively long leads, and I remembered some instances in which lead length had caused trouble. The cure was to cut the leads down to about 5" so that they could not be too near the sync stage.

Intermittents from Steady-State Defects

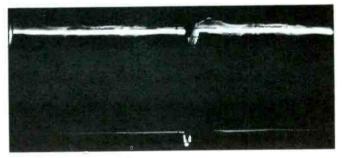
The level of the blackest picture signals in the composite video-sync signal is the primary factor explaining intermittent sync troubles when the sync or video systems have con-



(A) Normal



(B) Abnormal Fig. 6. Waveforms obtained at grid of horizontal AFC.



(A) Abnormal plate signal



(B) Normal plate signalFig. 7. Trouble caused by 6BA8 tube.

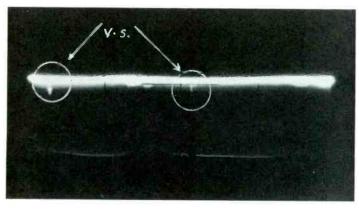
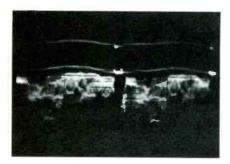
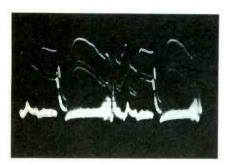


Fig. 8. Minute vertical-sync pulses.

stant defects. If the blackest picture signals are slightly below the blanking level (Fig. 9A), sync trouble will not result even though the video amplifier or the sync system has minor defects. However, when the blackest picture signals attain the blanking level, as they do during commercials or other high-contrast transmission (Fig. 9B), the highlevel picture signals ride through the separator and precede the vertical sync pulse; vertical rolling or jitter results: Even when the separator is stripping below its proper level, the signals in Fig. 9A will be correctly synced, but the signals in Fig. 9B will result in a contaminated sync output (Fig. 9C). This contamination can cause horizontal sync trouble. Signal distortion in some video-amplifier pentodes produces intermittent sync when the com-



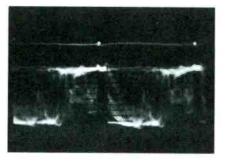
(A) No effect on sync



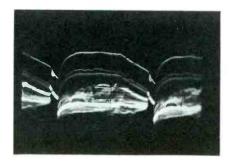
posite signal has high picture levels. With a normal composite-video signal, this same distortion has no effect.

Intermittent vertical flipping (Fig. 10) for short periods is frequently associated with steady distortion produced entirely within the videoamplifier stages. Figs. 11A, 11B, and 11C, found on the plates of 6BA8, 17BY7, and 6BH8, respectively, are some examples. The picture was steady when these signals were taken, but when the black picture signals reached the blanking level (Fig. 11D) the flipping burst occurred.

The conditions shown in Fig. 11 are limited to video-amplifier stages that are RC coupled to the video detector and thus have high input impedances. The effects of these distortions on the sync-separator



(B) Video in syncFig. 9. How composite signal can vary.



(B) Plate of 12BY7 Fig. 11. Distorted composite video.

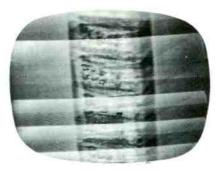
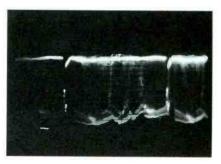


Fig. 10. Intermittent vertical sync.

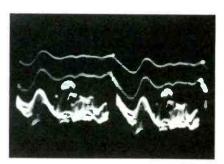
signals are widely variable. In some cases, the vertical-sync pulse groups are widened as in Fig. 7A; at other times, the dropout distortion shown in Fig. 12A takes place; and then again, the vertical sync may be attenuated as in Fig. 12B.

More Video-Signal Distortion

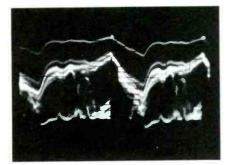
Another class of intermittent vertical flipping with extremely critical sync has been traced to videosignal distortions. In one Motorola receiver, an open filter capacitor at the screen of a 12BY7 video amplifier caused the distortion shown in Fig. 13A. The condition in Fig. 13B was due to an open filter capacitor on the 150-volt line in an older Philco. An open filter capacitor on the B+ line produced the symptom in Fig. 13C in a Sears set. Notice that the sync-blanking levels and the



(C) Distorted sync pulses



(C) Plate of 6AU8



(D) Caused vertical roll

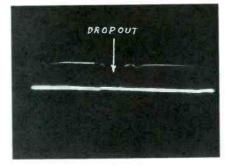
Fig. 11. Distorted composite video.

sync-tip levels do not contain an abnormal amount of ripple. The hash, or unwanted signal, directly on the bad capacitors is illustrated in Fig. 14A for the Motorola and Fig. 14B for the Philco (the Sears filter had a minimum amount of hash).

This form of distortion is lowfrequency loss in the video signal. It is most evident in the portion of the composite video signal most closely associated with low frequency. Other forms of low-frequency troubles have been discussed in a previous PF REPORTER article on AGC filters.

Other Troubles

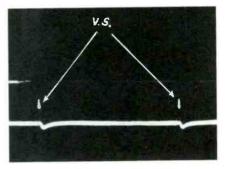
Other sync troubles are encoun-



(A) Complete dropout Fig. 12. Effect of distortion on sync.

tered frequently. One is the timelapse sync trouble which occurs only after a normally operating receiver runs for a period of time. Since this trouble shows up virtually every time, it isn't precisely correct to consider it as an intermittent.

Another form of vertical instability is caused by the minute difference between the vertical-frequency rate in color and black and white. Since these troubles occur only in monochrome receivers during color reception, they are sometimes considered as intermittent troubles. This condition arises from interaction between the 60-Hertz powerline frequency and 59.94-Hertz vertical-sync frequency with color. This trouble is common in some designs

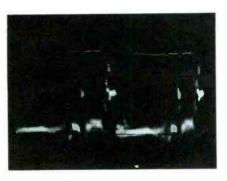


(B) Amplitude reduced

and has led to the issuance of a great many production-change bulletins by receiver makers.

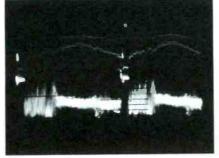
Conclusion

The one point that cannot be emphasized too strongly in troubleshooting intermittent sync troubles is that here is a place to use an oscilloscope. Many intermittent sync troubles are due to defects that cause steady signal distortion. Remember, only a scope is capable of indicating the minute distortions that are responsible for the picture symptoms. The most critical examination may sometimes involve expanding the signal traces. The practice of examining video sync signals critically will pay off in time saved.▲

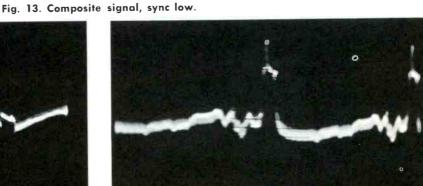


(A) Motorola receiver

(A) In Motorola set



(B) Philco receiver (C) Sears receiver



(B) In Philco set



REMOVING THE



In color television, removing the "barber pole" is slightly more difficult than removing the pole at a haircut shop. A loss of color sync often presents a challenge to the best service technicians. However, several factors can make the job a bit easier. Experience, of course, is most valuable in these circuits. Knowing how the circuit functions is even more valuable.

Reactance Controlled Oscillator

The majority of color instruments presently in the field have similar color-sync stages. These circuits are centered around a reactancecontrolled 3.58-MHz oscillator. Rapid repair of these stages follows naturally if: 1) the basic circuit functions are understood, (2) symptoms produced when color sync fails can be interpreted, and (3) a systematic procedure is developed to quickly isolate what circuit area is at fault.

The block diagram in Fig. 1 illustrates the most popular circuit configuration. The shaded blocks indicate those stages to concentrate on when failure of color sync causes multicolored stripes in the picture. To narrow our objective, let's assume the loss of color sync is a bench job; normal field service such as tube replacement has been to no avail. On the bench, we'll have readily available service data, color generator, scope, etc.

Color sync defects can be clas-

sified according to the results displayed on the screen. A complete loss of color sync—color bars on the screen, broken into a number

the screen, broken into a number of horizontal bands—is produced when the 3.58-MHz oscillator operates at the wrong frequency. Fig. 2 illustrates a small error; Fig. 3 shows a large frequency error.

Weak color sync is usually evidenced by good sync on strong sigsignals. In other words, color sync is too critical. For example, sync slips out of lock when the fine tuning is only slightly mistuned from crossover; or color is locked on one station, out of sync on another. This problem, like weak horizontal sync in black-and-white circuitry, is perhaps the most difficult to solve. Successful troubleshooting often involves detailed circuit analysis using a scope and bar generator.

Incorrect colors, indicating improper phase relationships, may also be classified as color sync trouble. The color may be locked, but wrong hues are produced on the screen. This is because the tint control circuitry is generally located in the color sync section.

Other symptoms can be produced by failure in the 3.58-MHz oscillator circuitry. An inoperative oscillator causes a complete loss of color, for example. However, to narrow our thoughts, let's concentrate mainly on one of the three symptoms mentioned above—complete loss of sync. Fig. 4 is a simplified breakdown of a phase detector/reactance controlled oscillator. Note the basic layout and learn to establish the major adjustments and operating particulars of this system. Then, regardless of the color instrument involved, a quick glance at the service schematic will show the key points for any similar circuitry.

Four major sections combine to form our sync system: the burst amplifier, phase detector, reactance control tube, and 3.58-MHz oscillator. In this closed loop system, the correct operation of the oscillator depends on correct operation of all four sections. Table I shows how these stages are similar to b-w stages.

Complete Loss of Sync

The presence of fully saturated color on the screen, even though out of sync, helps pinpoint the defective stage. Our first suspicion and first tests—should be in the reactance tube and oscillator circuits. The burst amplifier and phase detector are secondary choices.



Fig. 2. Keyed-rainbow pattern with a small color sync error.

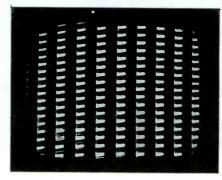


Fig. 3. Large sync error.

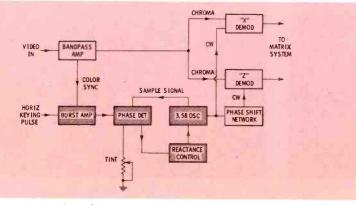


Fig. 1. Block diagram shows path of the burst signal.

Here's the reasons:

1. High level color reaching the picture tube tells us the killer stage is cut off, permitting the bandpass stage to pass chroma to the color demodulators.

2. Killer cut-off voltage is obtained via circuits associated with the output of the burst amplifier, thus, separated color burst (sync) is probably reaching the phase detector. If sufficient burst were now present at the output of the amplifier stage, the result would be a no-color condition. It's a simple task to confirm our analysis.

Reactance-Oscillator Test

With a color bar signal applied, ground test point TP1 (using a short clip lead), and observe the screen pattern. This action virtually removes the burst amplifier and phase detector stages from influencing the frequency of the oscillator. As we'll see presently, the results of this simple test are two-fold; tying trouble to the reactance-oscillator or the burst-detector sections.

Grounding TP1 tells us if the

Table I

Color Sync	Similar
Stage	B-w Stage
Burst Amp	Sync Sep. Horiz AFC Diodes
Phase Det. Reactance Tu	be Horiz Osc Control
3.58-MHz Os	

3.58-MHz oscillator is capable of running at the correct frequency. Observe the screen for one of the following:

1. If the color bars become vertical (normally positioned) on the screen, become stationary, or slowly drift across the screen, then the oscillator and reactance stage are in good shape.

2. If the number of horizontal color bands (large frequency error, initially) decreases in number to three or four bands of solid colors, then the oscillator is probably okay; under these conditions, the initial loss of sync was not directly caused by the oscillator. Here, it's a good idea to make a slight adjustment of the oscillator frequency coil to

obtain "zero" beat — color bars slowly floating across the screen, setting the free running frequency as close to 3.58 MHz as possible.

3. If the color pattern remains out of sync when TP1 is grounded, then look for trouble directly in the reactance or oscillator stages. Voltage, resistance, and substitution tests will be necessary. Leave the test point grounded during trouble-shooting procedures, to keep the oscillator isolated from the other sections.

If color bars could be synced (results 1 or 2), then trouble is evident in the burst or phase detector stages. The following test helps to further isolate to a single stage.

Remove the clip lead from TP1; color sync will again be lost, because a defective phase detector or burst amplifier is causing a wrong correction voltage to be developed at the output of the phase detector (TP1). Under normal conditions, the phase detector has a balanced output (both diodes conduct equally) and the voltages at the junction of R1-R2 yield zero correction voltage. If an unbalanced condition develops, for any reason, the DC error voltage fed to the reactance tube pulls the oscillator off correct frequency, causing a loss of color sync. The amount of DC voltage error determines the frequency error of the oscillator.

The problem now is to determine which stage—burst amplifier or phase detector—is causing the error voltage to be developed. The next test is to check the phase detector.

Phase Detector Test

Connect a short clip lead to the grid of the burst amplifier and ground. This action removes the incoming burst signal, permitting us to check the balance of the phase detector. Conduction of the detector diodes, under these conditions, is influenced only by the sample CW signal from the oscillator. Both diodes receive the same signal (anode and cathode tied together), and conduction of both diode sections should be equal, but opposite. Using your VTVM, perform the following:

1. At the output side of the phase detector, measure the anode and cathode voltages. If the voltages are equal and opposite the phase detector is balanced. Hence, the trouble probably exists in the burst amplifier circuit, which includes the secondary of the burst transformer T1.

2. If the voltages on the anode and cathode are unbalanced, the fault lies in the phase detector.

3. If the improper amount of equal voltage is developed, or if zero voltage is developed, check for loss of the sample signal from the oscillator. (The +28 and -28voltages shown in Fig. 4 are typical with no burst input. For the correct values for any particular chassis the proper PHOTOFACT folder should be consulted. Possible causes of an unbalanced phase de-• Please turn to page 53

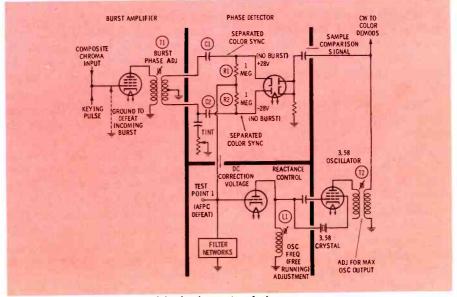


Fig. 4. Simplified schematic of chroma sync stages.

THE CASE OF THE

Bifilar Balun

by Jack Darr

The lightning flashed; the thunder roared; and my phone rang. "My TV set's out completely! No picture, just a lot of snow. I've got guests coming tonight to see a new show in color, and I want it fixed! Can you come right over?"

"Yes Ma'am!"

This was my best customer (a big antenna installation, one color and three b-w sets, stereo, and more stuff that you can shake a stick at). Besides this, she was a widow, and rich, and a very nice little old lady. Yes, ma'am; I'd be there.

It was a Zenith 25MC32 color chassis, and the picture was snowy; in fact, there was darn little picture at all. Remembering the storm, I asked "Was there a big flash of lightning around here this morning early?"

"Oh, my yes! I thought it'd break every window in the house. Terrible!"

Oh, oh. Take the back off and look at the tuner. Yep. There it was, or rather, there's where it used to be. The balun coil was a disaster area; little pieces of wire everywhere. Everywhere but where they ought to be. So, I pulled the tuner, and lit out for the shop (I love separate tuners!)

I called the distributor. Tragedy! No replacement coils, and none expected for a week! No replacement tuners, either, so I was desperate. I hung up and glared at it. Well, genius, don't just sit there looking stupid. Fix it!

Thank goodness, this wasn't one of those sub-miniature types buried deeper than a time capsule inside the tuner. I dug out the PHOTOFACT Folder (757-4) and looked at it. Didn't look too difficult (Fig. 1). There was enough left of it so that I could tell that the balun had been a bifilar-wound type; the original had green and white wires so that you could tell which was which. (There was enough of the insulation left so that you could see this.) There were two coils, spaced about 1/4-inch apart, on a nice big ferrite core mounted in snap-in brackets, and on top of the tuner, too. So, I made a rough sketch of the core, capacitors, and the little chokes in the FM traps, just for luck, and went looking for wire to rewind it with.

Hmm. No wire. My "replacement stock" of coil-rewinding wire was an ancient horizontal oscillator coil that I'd had for years, but this was too little. I needed a wire that would be about the same size as the original, so that I could hold the spacing, and so on. Back to the junk room.

About thirty minutes later, I came out dusty but happy, with

the field-coil of a 6-volt auto-radio speaker that had been hanging on the wall for maybe 15 years or so. Just right. Nice enameled wire about #24 or #26. This would let me rewind them and come out with about the same turn spacing that the original balun had before the catastrophe.

So, out with the core. The loupe showed me that there had been 6 turns in each coil, so this was nice: I wouldn't even have to take my shoes off to count up that many. Both coils were wound in the same direction; that was good. I scraped off what was left of the original and put a band of nail polish on one end so that I could tell which way to wind. (Turned out this wasn't necessary, but it was a nice thought anyhow.)

Hoo, boy. Well, start. Don't just sit there. I cut off two pieces of the field-coil wire about 12" long, and doubled them. Smoothing them out so that they laid side by side, I wound on 6 turns and twisted the ends to hold it. When I picked up the form to wind the second coil on, the first one promptly fell off and rolled under the bench. The heck with it; I wound on another one. At my age, this was a lot easier than stooping over that far. (and getting back up again-always the hardest part). With two coils on the form, I squirted the thing with

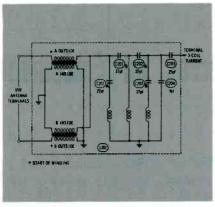


Fig. 1. Bifilar balun coil and FM trap with dual capacitors.

some silicone-resin lacquer to hold them in place, and went to get a cup of coffee while it dried. I knew if I stayed there, I'd start before the paint dried.

When I got back, it was dry. The coils were still loose, but the paint was dry. Oh, well. Now to hook them up. There are four coils, so I have 8 leads. To help hold down the confusion as much as possible, I named each one. "A-outside" (meaning the outside wire on one coil, at what was the outside end with the coil mounted in the clips) was also called the "start" end of the winding, and I put a little dot on the schematic to identify it. (Actually, it doesn't make any difference which end you call the start, as long as you can tell which coil is which. See Fig. 1 again.)

Now, I clipped off the ends of the coil wires, and tinned each one for about 1/4". Sucking an unfortunate fingertip that had somehow managed to get caught in the machinery, I spread them out, and clipped the coil into the brackets (Fig. 2.) Next, I sorted out the ends of each coil with the ohmmeter, and twisted their ends together so that I could tell which was which. Next, hook them up. Well, gotta start somewhere, so take the one thing you're sure of-the antenna terminals on the top. So, I hooked up "A-outside" to that one, using the "Start" lead. Now, where the heck does the other end go. Oh, I see; to the junction of two little capacitors. Hmmm. Only four capacitors in there, and I see 6 on the schematic. Get the loupe again! Oh. Two of them are duals; a close look shows that two of the little blobs have three leads each (C201 and C202, dual 27-pf ceramics).

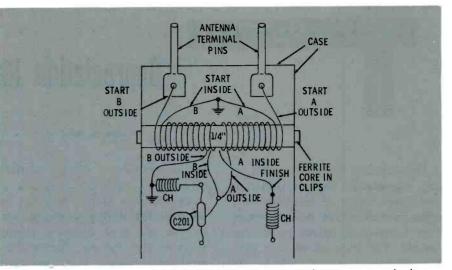


Fig. 2. Physical layout of the balun coil, FM trap, and antenna terminals.

There you are, you little dickens: A solder eyelet on the board, with a capacitor-lead and a lead from one of the trap coils in it. Up with the "tiny-tip" iron and stick the other end of "C-outside" there. That did it. Next, the other winding of coil A goes to—ground? Yep. OK. Now, the other end goes to—ground again? Yep. That's what I like about balun coils. They're such nice illogical-looking things. Now, for coil B.

I take the outside wire of "B", the start, and hook this to the other antenna terminal. The other end goes to ground again. Plenty of little ground-terminals around on the terminal board. On the inside wire of "B", the start end goes to ground, just for a change. There happens to be a ground terminal right between the antenna terminals (as you can see in Fig. 2) and this was where they'd been, so I stuck 'em back. The other end of "B-inside" goes to the junction of the capacitor and the end of the other coil. Nice; breaks the monotony of hooking everything to ground. (I was beginning to wonder if there were any hot wires on the thing) So, I tack this down, and somewhat to my surprise, I'm out of wires. That's all of them. Now, where was I? (Sitting here looking confused, that's where.)

Well, trace them all out again, just for luck. OK, looks good. Check the tubes, and look the tuner over carefully for any sign of other damage. Nope, can't see any. Now, we're off to see the widow. Arriving at her house with the tuner tucked under my arm and a smug look on my face, I tell her, "Well, I think I've got it fixed. It was a little tricky, but I did it!" With this modest remark, she let me in, and I reinstalled the tuner.

Crawling out from behind the set and getting up (with some difficulty), I turned it on. Well! Low channel stations pretty good. Nice color, not too much snow, and I begin to feel pretty smart. Now, let's try the high channels. Woops! Frankly, they're what I'd call "lousy." Back to the low channels. Come to think of it, they don't look as clear as they did when I put the tuner in. Oh, oh.

So, acting on the old motto, "When trouble arises, look at the last thing you put in, stupid", I crawl around behind the set and look carefully at the balun again. An ohmmeter won't do a lot of good with all of those ground connections, so it's up to the calibrated eyeball. Did I bring the schematic with me? Certainly not; are you kidding? So, I sat there on the floor with my legs under the cabinet (only position you can get at the tuner) and went into a trance.

Hey! Wait one. The hot wire from the outside coil on "A" and the inside one on "B" should go to a dual capacitor, but not to a coil! Yet, I could see it firmly soldered to the junction of a coil and capacitor! Out to the truck and get the soldering iron. Unsolder the wires and pull them carefully out with the little dykes I carry in my pocket (you didn't think I'd bring a pair of long-noses, did you?). I start looking. Right next to it is

• Please turn to page 52



Depreciation Reserves for Equipment

by Raiph H. Butz

Depreciating tangible assets test equipment, service vehicles, and other forms of machinery—is often considered as nothing more than a device to reduce an income tax bill. The value of an asset usually declines in some ratio to the amount of depreciation allowed for tax purposes.

The U. S. Treasury Department recognizes the economic necessity of depreciation allowance. Therefore, the tax code permits a business to establish depreciation reserves during the useful life of an asset, so that its cost may be recovered by the time its profitable use has been exhausted.

It is probable that the liberal depreciation allowances permitted under the 1962 tax law will not continue to be available for an unlimited period of years. When it is no longer economically feasible to encourage business expansion with tax-saving incentives, and when additional revenue is needed, it is possible that such incentives as the investment tax credit on equipment purchases and the additional first year depreciation allowance will be rescinded.

In addition to the 7% investment tax credit on equipment purchases and the additional 20% first year depreciation allowance, some businesses have also used one of the various forms of accelerated depreciation to obtain maximum deductions for tax purposes.

What is depreciation? It is the steady decline in value of assets used to produce business income. The major factors responsible for the decline in value are wear and tear, obsolescence, and depletion. Under certain conditions, the rate of depreciation may be geared to production instead of a specific period of time. This would apply to manufacturing or processing operations where equipment would be considered as having a useful life to produce a certain number of units.

Depreciation credits may not be accumulated during years of low income and later deducted during years of high income. Depreciation must be deducted each year, and the deduction must represent a reasonable allowance allocated annually over a period of years designated as "useful life of the equipment." Also, the manufacture or processing of a certain number of units can be used as a basis for the rate of depreciation.

"Useful life" and "physical life" of equipment may differ to a large degree. When you estimate the useful life of equipment you are planning to use it only as long as it is considered efficient or profitable for your business.

What might be designated as the useful life of an item in one business might not be applicable in another business. This fact is recognized by the U.S. Treasury Department as indicated in the following statement: "The useful life of any item depends upon such things as the frequency with which you use it, its age when you acquire it, your policy as to repairs, renewals, replacements, the climate in which it is used. . . the normal progress of the arts, economic changes, inventions and other developments within the industry, trade or business."

Property with a useful life of one year or less—such as small tools, accessories, books, etc.—is not depreciable, but is charged to expenses in the year purchased. Tangible assets with a useful life of more than one year are usually depreciated by using the straightline method—deducting an equal amount each year. Other forms are authorized by the Treasury Department and may be used under certain conditions.

Tangible property with useful

life of six years or more, used to produce business income, may qualify for the additional 20% first year depreciation allowance in addition to the regular depreciation. If the property has useful life of more than four years, it may also qualify for the investment tax credit on purchased equipment.

Brown purchased equipment on January 1, 1966, for \$6,000, estimating its useful life as five years, with \$500 salvage value at the end of that period. When the salvage value of an item is 10% or less of the original cost, it may be disregarded for purposes of computing depreciation. Since the equipment has a useful life of less than six years, he is not entitled to the addtional 20% first year depreciation allowance. The depreciation schedule, using the straight-line method, is as follows:

Cost of equipment

cost of equipment		
January 1, 1966		\$6,000
1966 Depreciation		
20%	\$1,200	
1967 Depreciation		
20%	1,200	
1968 Depreciation		
20%	1,200	
1969 Depreciation		
20%	1,200	
1970 Depreciation		
20%	1,200	6,000

Assuming that Brown had estimated the useful life of the equipment as six years, enabling him to qualify for the additional 20% first year depreciation allowance, the straight-line depreciation schedule would change to: Cost of equipment

January 1, 1966	\$6,000
Less 20% additional	
1st year depreciation	1,200
	1 0 0 0

4,800

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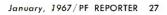
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TV servicing is a highly competitive occupation. Efficiency in our job requires a good understanding of Ohm's law. Otherwise, we cannot analyze our measurements to localize defective receiver sections and to pinpoint faulty components. If we understand Ohm's law, we have the most powerful tool avail-

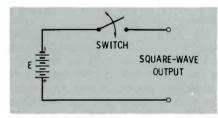


Fig. 1. A simple square-wave generator using a mercury switch.

able to tackle any electronic circuit. Most of us have a good understanding of Ohm's law for DC, but are a bit hazy concerning Ohm's law for AC. Beginning technicians tend to feel that AC voltage and current are too difficult to understand, but this is not so. In fact, when the basic idea of an AC voltage or current is pointed out, the seeming difficulties disappear immediately.

From DC to AC

There is a simple relation between DC and AC that can be shown in various ways. One way is depicted in Fig. 1. If the switch is left closed, we have a DC source. On the other hand, if the switch is opened and closed rapidly, we have a square-wave source. A squarewave is one basic type of AC. Note that in the past, mercury switches were widely used to change a DC source into a square-wave source as seen in Fig. 1. Today, we use vacuum tubes and transistors instead of mercury switches, but the operating principle is the same.

Next, observe Fig. 2. A series RC circuit is connected to a squarewave generator. This is an integrating circuit, with the scope connected across the capacitor. The voltage across the capacitor has a waveform that can be plotted by the universal time constant chart. In other words, the voltage rises across the capacitor as shown by curve A in Fig. 2B. At the same time, the current flow into the capacitor is shown by curve B. There is an important basic fact to be observed here: Note that the current flow into the capacitor has its maximum value when the current flow into the capacitor is zero. On the other hand, the voltage across the capacitor has its maximum value when the current flow into the capacitor is zero. This

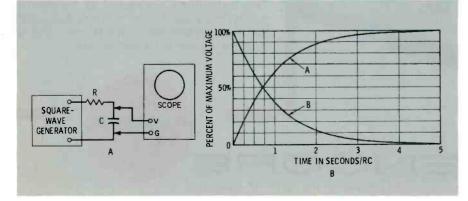


Fig. 2. Series RC circuit and chart showing voltage and current.

is just another way of saying that the current leads the voltage in a capacitor. Beginning technicians often have a hard time understanding how current flow can be maximum when the voltage is zero. However, the mystery is quickly cleared up by observing the circuit action in Fig. 2A. The voltage rises gradually across the capacitor because Q = CE, where Q is the quantity of electricity in the capacitor. Or, we can say E = Q/C. Since Q = It, where t is the time, it is obvious that E will be zero when t is zero, and that E will increase as t increases. However, E does not increase beyond a certain point, because I is decreasing as t increases.

Sine-Wave Voltage and Current

Now, let us use an audio oscillator instead of a square-wave generator to drive the RC series circuit. Fig. 3 shows the circuit arrangement. The voltage across the capacitor and the voltage across the resistor can be seen in Fig. 3B. Observe that the voltage across the resistor leads the voltage across the capacitor by 90°. This is the same as saying that the current in the capacitor leads the voltage across the capacitor by 90°. Why? It is because the voltage across a re-

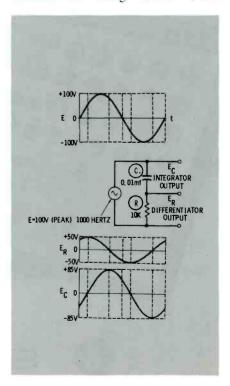


Fig. 3. Voltages across C and R of RC circuit with sine-wave input.

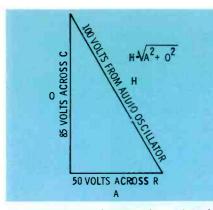


Fig. 4. Using right triangle and Pythagorean formula to add RC voltages.

sistor is in phase with the current through the resistor.

You will also observe a surprising fact in Fig. 3. Here we have a generator that drives the RC circuit with 100 volts p-p. Note carefully that the peak voltage across the capacitor is 85 volts, and the peak voltage across the resistor is 50 volts. In other words, the simple sum of $E_{\rm R}$ and $E_{\rm C}$ is 135 volts. How can the sum of the circuit voltages be greater than the source voltage? The mystery disappears when we observe that E_c and E_R are 90° out of phase. Therefore, we must add Ec and ER vectorily at right angles, as shown in Fig. 4. This is the basic difference between AC circuits and DC circuits. We have to consider phase relationships when we analyze AC circuits.

Voltages Around a Square-Wave Circuit

We saw in Fig. 4 how voltages add up around a sine-wave circuit. The vectors in Fig. 4 are merely a shorthand method of adding up the sine-wave voltages depicted in

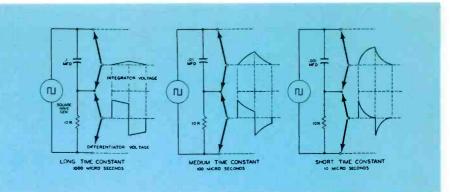


Fig. 5. Waveforms in RC circuits having different time constants.

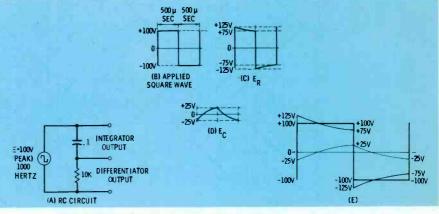


Fig. 6. Sum of voltages across R and C is equal to the applied square-wave voltage.

Fig. 3. We add the peak instantaneous voltages, taking the phases of the sine-waves into account. Next, we shall see that the voltages around a square-wave circuit also add up in basically the same manner. Fig. 5 shows voltage waveforms in series RC circuits that have different time constants. In every case, the sum of the instantaneous voltages across the resistor and capacitor add up to the applied square-wave voltage.

An RC circuit with all voltage values indicated is shown in Fig. 6. We expect that at each instant, the

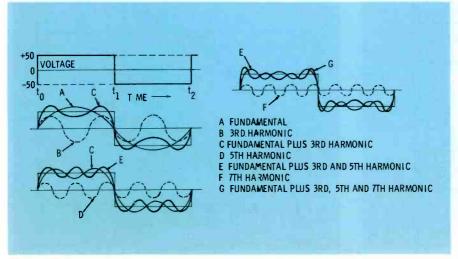


Fig. 7. Infinite number of sine waves in symmetrical square-wave.

sum of the voltages across R and C will be equal to the applied square-wave voltage. Fig. 6E shows that this is so. Fig. 6 is a bit more complicated than Fig. 4 because a square-wave cannot be drawn as a vector. Hence, we must add the entire waveforms across R and C to show that their sum at any instant is equal to the applied squarewave voltage. Note that both Fig. 6 and Fig. 4 are examples of Kirchhoff's law, which states that the sum of the voltage drops around a complete circuit is equal to the applied voltage.

It is sometimes asked why a square-wave cannot be represented by a vector. The answer to this question is shown in Fig. 7. Since a square-wave consists of a large

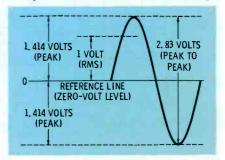


Fig. 8. Relation of peak, RMS, and peak-to-peak voltage values.

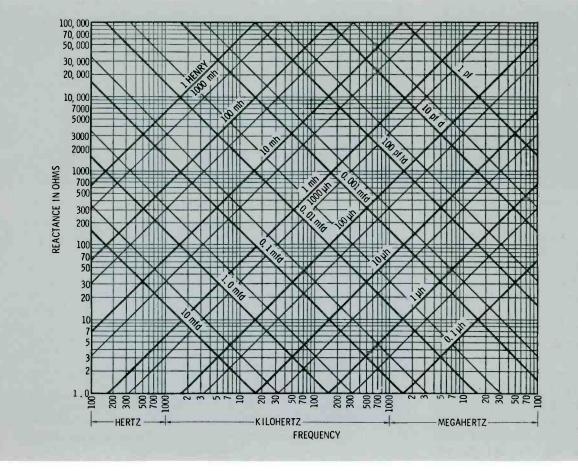


Fig. 10. Reactance versus frequency for inductors and capacitors.

number of sine waves (an infinite number of sine waves in the ideal case), and each sine wave corresponds to a vector, we would have to draw an infinite number of vectors to represent an ideal squarewave. Since this is impractical, we use the universal time constant chart (Fig. 2B), instead. The waveforms in Fig. 6 have the same curvature (exponential shape) as the universal waveforms in Fig. 2B. The curvature may seem to be different simply because the various illustrations are drawn to different scales. Ohm's Law for a Capacitor

We know that Ohm's law for a resistor states that I=E/R. This law holds true whether we are speaking of DC values, peak AC values, peak-to-peak AC values, or rms values. It is only necessary that

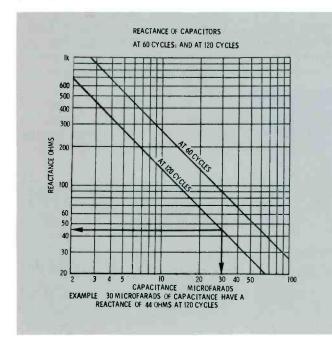


Fig. 9. Reactance of capacitors at 60 and 120 hertz.

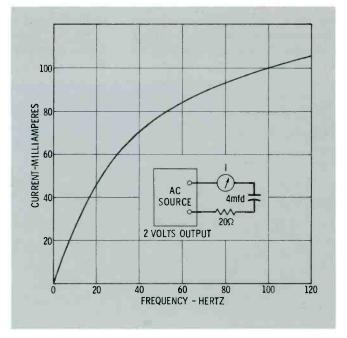
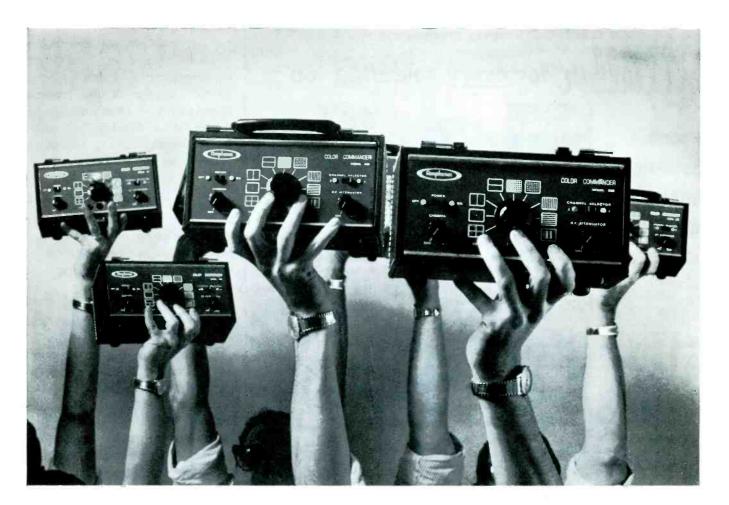


Fig. 11. Current versus frequency in a series RC circuit.



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we be consistent in the equation when we choose values. We can easily convert from one value to another by use of the relations shown in Fig. 8. Ohm's law for capacitors states that $I=E/X_c$, where X_c is the capacitive reactance in ohms. From the standpoint of arith-

metic,
$$X_c = \frac{1}{(2 \pi fC)}$$
, where X_c is

the reactance in ohms, π is equal to 3.1416, f is the frequency in hertz, and C is the capacitance in farads.

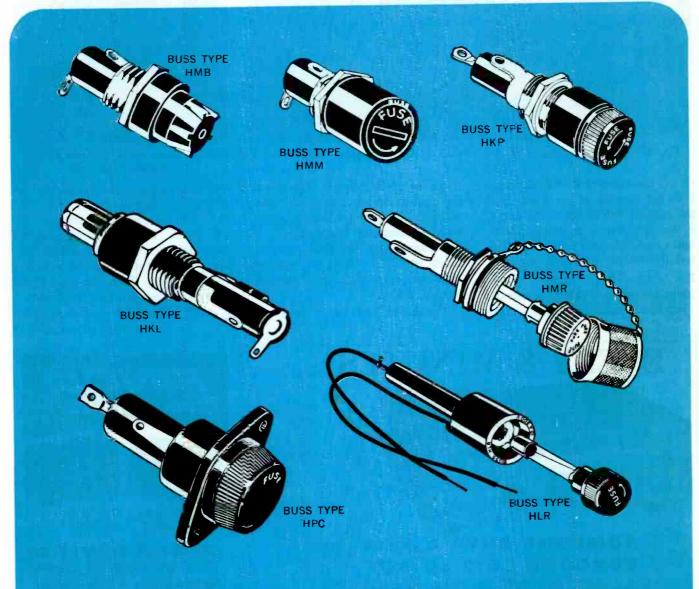
Fig. 9 shows how the reactance varies with capacitance at 60 and 120 hertz. At any given value of capacitance, the reactance at 60 hertz is double that at 120 hertz. At any given frequency, the reactance is halved when the capacitance is doubled. Of course, these facts are obvious from inspection of Ohm's law: $I = E/X_c$. Note that resistance is a voltage-current ratio: $\mathbf{R} = \mathbf{E}/\mathbf{I}$. In the same manner, reactance is a voltage-current ratio: $X_{c} = E/I$. Technicians are busy people-they use timesavers whenever possible. Therefore, you should tack Fig. 9 over your service bench. It will throw immediate light upon many power-supply filter problems.

Of course, troubleshooting is often concerned with frequencies other than 60 to 120 hertz. The chart in Fig. 10 shows the reactance of just about any capacitor or inductor at any frequency that you might be concerned with in radio, audio, and TV trouble-shooting. Note that Ohm's law for inductors is basically the same as for capacitors. In other words, $I=E/X_L$, where X_L , the inductive reactance in ohms, is equal to 2π fL.

Working With Impedance

When a circuit has both resistance and reactance, it is said to have a certain impedance, which is measured in ohms. Fig. 11 shows a series RC circuit, and how the current flow varies with frequency.

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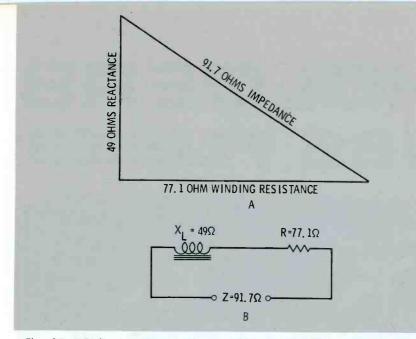


Fig. 12. Winding resistance of coil added to reactance gives impedance.

Ohm's law states that I=E/Z, where I is the current in amperes, E is the applied voltage in volts, and Z is the impedance in ohms. Reactance and resistance are added vectorially (right triangle), so that $Z = \sqrt{R^2 + X^2}$. Observe Fig. 12. Inductors have winding resistance. Therefore, inductors offer impe-

dance in an AC circuit. The winding resistance acts as if an ideal inductor were connected in series with a resistance. Accordingly, if an inductor has 49 ohms of reactance and 77.1 ohms of winding resistance, the impedance of the coil is 91.7 ohms.

Capacitors have very little re-

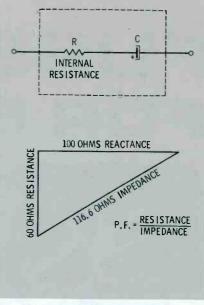


Fig. 13. Series resistance added to reactance of capacitor.

sistance, and in turn a capacitor is a reactor. A capacitor is not, in the normal meaning, an impedance. Of course, if an electrolytic capacitor becomes defective, it will develop leakage resistance or a highresistance film. Then, we must analyze the capacitor as an impedance. In Fig. 13A, the capacitor has • Please turn to page 55





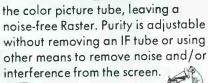


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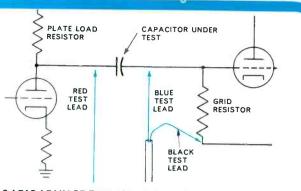
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BOOK REVIEW

Communications Electronics Circuits: J. J. DeFrance; Holt, Rinehart and Winston, New York, 1966; 548 pages, 6" x 9", cloth, \$9.50.

This book presents a detailed analysis of each of the principal circuits used in RF applications. Circuits are grouped according to purpose, such as RF voltage and amplifiers, oscillators, power modulators, and detectors. Complete chapters are devoted to combined circuits in transmitters and receivers. Frequency modulation is treated separately, as are transmission lines and antennas. Special chapters are given to resonant and coupled circuits, and to special transmitters.

Each circuit described is developed by mathematical analysis, and most circuits now in use are discussed. At the conclusion of each chapter are several pages of test questions and design problems. Their solutions require considerable review of each circuit and problem covered by the book.

This is a text on the subject and can be used as a handbook of RF design. The use of mathematics, extending to integration, is extensive, but does not preclude use of the volume without envolved calculations.



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Circle 16 on literature card January, 1967/PF REPORTER 37

Notes on Test Equipment

analysis of test instruments . . . operation . . . applications

by T. T. Jones

Mutual Conductance Tester

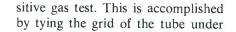
One of the more interesting of the many new test instruments is **Mercury's** new Model 2000 tube tester. This tester offers a variety of tests, and can test every known tube and transistor type, with the exception of the obsolete numbers such O1A, 27 etc.

The setup is very fast, thanks to the lever switches. Fig. 2 shows a portion of the set-up book, and shows, in the Gm-Em column, which levers must be thrown to each position. The levers are smooth and lightly detented as is the SELEC-TOR switch. The latter controls heater voltage.

Another time-saver in the set up is the use of multiple sockets. The heaters are permanently wired in, with a separate socket for each known heater connection.

Switches 1 through 10 connect the load and bias voltages to the corresponding tube pins. Switch 11 selects the proper bias voltage, and 13 selects the plate voltage. Switch 12 selects the heater return circuit.

In addition to the mutual conductance test, there is a very sen-



Mercury Model 2000 Speceifications Tests Performed

Tubes: Shorts (1 M Ω sensitivity), Gas (150 M Ω sensitivity), Emmision, Mutual Conductance (0-5000 μ mho, 0-25,000 μ mho sensitivity). Transistors: Leakage current, DC Beta (0-200) Types Tested: 5 and 7 pin Nuvistor, 7 pin

- 5 and 7 pin Nuvistor, 7 pin miniture, Octal, Loctal, Noval, Novar, Magnoval, Decal, 10pin Sylvania, and Compactron, in all present basing configurations. CRT's with optional adaptor.
- Power Requirements: 117 VAC.
- Size (HWD): $4\frac{1}{2}$ " x 18 $\frac{1}{4}$ " x 10 $\frac{3}{4}$ ". Weight:

93/4 pounds.

Price: \$99.95

1] I SHORTS G-1/# G-1 0-1

Fig. 2. Chart shows typical set-up data.



Fig. 1. Lever switches are featured in the Model 2000.

test to a VTVM circuit. The sensitivity of the gas test is $150 \text{ M}\Omega$.

The Model 2000 can also be used as an emission tester for those tubes which are impractical to test by the Gm method, such as the 6BK4, rectifiers, etc.

The transistor test is a very simple one, measuring leakage between collector and base, then applying forward bias for the beta measurement. The tests are made with a DC supply voltage.

A very helpful added feature is the line of pin straighteners just to the left of the meter. In addition to the usual 7-and 9-pin straighteners Mercury has included straighteners for Nuvistors, Novars, and compactrons. Most testers don't have even the 7- and 9-pin straighteners.

The unit is housed in a suitcasestyle wood case, with leatherette covering. There is a large cable compartment which can hold the accessory CRT testing cable. This cable is available at extra cost (\$12.45) and enables the Model 2000 to test all CRT's including color types.

> For further information circle 52 on literature card

RF Signal Generator

Pictured in Fig. 3 is **RCA's** new WR-50B signal generator. At first glance, it appears to be an update o fthe older WR50A, which we analyzed in the June 1964 issue of PF REPORTER. After a look at the schematic, and a peek inside, about the only thing we could find that hadn't been radically changed were the knobs.

They also retained the same fine tank circuit components. As we noted before, the stability and calibration accuracy of the WR50 is exceptional for an instrument of this price range. We checked the drift on the WR50B, and from a cold start to an hour later, frequency moved less than 50 kHz at 30 MHz. Calibration accuracy is about as close as you can eyeball it at that frequency. We measured and found the accuracy was better than 1% on all bands.

Among the operational features new to the WR-50B is an attenuator for the Xtal oscillator. This in-



Fig. 3. RF Signal/Sweep Generator.

RCA WR-50B **Specifications** Variable Oscillator Frequency: 85kHz-40MHz in 6 bands. (Has strong second harmonic to 80MHz). Output: 50,000 µV minimum on fundamental. Calibration Accuracy: $\pm 2\%$. Internal Modulator: 400Hz, 0-30%. External Modulation: 15-kHz maximum frequency, 10 volts required for 30% modulation. Sweep Oscillator Center Frequency: 455kHz and 10.7MHz in 2 fixed bands. Sweep Width: Approximately 10% center frequency. Crystal Oscillator Frequency: 100kHz-15MHz. Output: 20.000 µV. Internal Modulation: approximately 20% Audio Output: 8VRMS across 15 K ohms. Attenuator VFO & Sweep: 10-1 switch plus variable. Xtal: Additional 7-1 step. Power Requirements: 105-130VAC 15 watts. Size (HWD): 73/4" x 53/8" x 43/4". Weight: 5 pounds. Price: \$65.

• Please turn to page 69



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Industrial Electronic NoteBook

by Ed Bukstein

The basic circuits employed in industrial electronics — amplifiers, rectifiers, oscillators, etc.—are essentially the same as their counterparts in other types of equipment. The radio-TV technician can therefore examine an industrial diagram and recognize these familiar circuits in unfamiliar surroundings. Certain features of the industrial circuit, however, may seem strange or mysterious.

Dual Primary Transformer

The radio-TV technician might be very much surprised, for example, when he discovers that power transformers in industrial equipment often have dual primary windings (Fig. 1). Actually there is nothing mysterious about the dual-primary transformer; it is employed to permit operation from either 115- or 230-volt power sources, both of which are generally available in industrial surroundings. The two 115volt windings are connected in parallel for 115-volt operation, and in series for 230-volt use. The connections are made so that the magnetic fields produced by the two

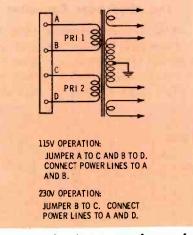


Fig. 1. Dual primary transformer for 115- and 230-volt operation.

windings aid rather than oppose each other.

Manual/Automatic Control Circuits

The typical on-off device in radio and TV sets is a SPST switch in one side of the power cord. Industrial equipment, however, often employ a more elaborate turn-on and turn-off circuit. For example, automatic shutoff may be required in the event of inadquate flow in a water-cooling system, excessive temperature of critical components, opening of an access door, etc.

In the typical arrangement shown in Fig. 2, the start cycle is initiated by means of the ON push button, which completes the circuit to the coil of line contractor CR1. Normally open contacts CR1-B and CR1-C now close, and line voltage is applied to the equipment to be controlled-an X-ray machine, an induction heater, a motor-control unit, etc. In addition, a normally open contact, CR1-A closes in parallel with the pushbutton. The operator can now release the push button, and the contacts remain closed.

The power can be turned off either manually or automatically. Manual turn-off is accomplished by pressing the OFF button; this deenergizes the coil of CR1, and all of the CR1 contacts return to their normal (open) positions. Automatic turn-off is controlled by the normally closed contacts in series with the OFF button. These contacts are components of protective devices such as overload relays, flow switches, temperature limit switches, door interlock switches, etc. If any of these normally closed contacts should open, contractor CR1 is deenergized and the equipment is disconnected from the line.

Anti-Chatter Relay Circuit

Thyratrons are frequently employed as relay-controlling components. In the circuit shown in Fig. 3, the relay coil is connected in the plate circuit of the thyratron, and the grid is held negative by a control signal. The relay remains de-energized as long as the control signal is present. The control signal can be supplied by a phototube, temperature-sensing circuit, limit switch, etc. When the control signal is reduced or removed, the thyratron ionizes and the relay becomes energized.

Because the grid of a thyratron loses its effectiveness as a control element after the gas is ionized, some other means must be employed to de-ionize the thyratron. An AC plate supply performs this function in Fig. 3. The thyratron deionizes during each negative alternation of plate voltage, permitting the grid to regain its control function during the first negative alternation after the control signal has been restored to the grid.

During the interval when the con-

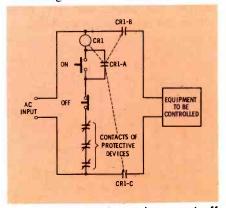
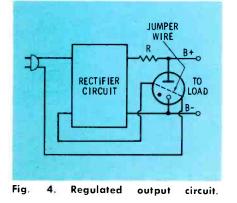


Fig. 2. Input circuit can be turned off automatically or manually.

trol signal is not present at the grid, the relay should remain energized. However, during each negative alternation of plate voltage, the relay will attempt to de-energize. This will cause the relay to chatter at the line frequency. To prevent this chatter, a capacitor is connected across the relay coil, as shown. This capacitor charges through the thyratron during the positive alternations of plate voltage. A resistor in series with the capacitor limits the surge current which might otherwise damage the thyratron. During the negative alternations of plate voltage, the discharge current of the capacitor keeps the relay energized. Therefore, the relay remains continuously energized during the interval when the control signal is not present at the grid.

Voltage Regulation

Voltage regulation in industrial circuits is frequently more critical than the regulation required in radio and TV circuits. This is particularly true of the high-sensitivity circuits used for industrial measurement and automatic control. A simple but effective regulator circuit (shown in Fig. 4.) employs a voltage regulator tube connected across the output of the power supply. The voltage regulator tube is a coldcathode, gas-filled diode. As is characteristic of gas-filled tubes, the voltage drop across the ionized gas remains constant. An increase of current flow through the tube will not produce an increase in the voltage dropped across the tube. The voltage across the tube remains constant because an increase of current is accompanied by a decrease in resistance of the ionized gas. Since the resistance decreases when current increases, the IR drop across the tube does not change.



Resistor R, in series with the regulator tube, "absorbs" the changes of supply voltage. If the line voltage should increase, for example, the regulator tube will draw more current. The voltage across R will increase, allowing the output voltage to remain relatively constant over a wide range of line voltage variations or load changes.

A jumper wire inside the regulator tube is connected between the two base pins. The purpose of this jumper is to protect the load circuit from the effects of excessive voltage. If the regulator tube is pulled out of its socket, during servicing procedures for example, less current will flow through series resistor R. Since the drop across R is now smaller, the output voltage increases and may damage the load. As shown in the diagram, the jumper wire in the tube is connected in series with the AC input. Therefore, removal of the tube from its socket will disconnect the input to the rectifier, preventing damage to the load.

Modulated-Light Photorelay

Photoelectric relays are used extensively in industrial applications, such as production counting, binlevel control, automatic packaging, etc. Since these circuits must operate in well-lighted industrial environments, they must be designed to "ignore" ambient light. In most applications this is not a serious problem because the control light source is located within a few feet of the phototube. The intensity of the source, as seen by the phototube, is therefore considerably greater than the intensity of the background illumination. For this reason, the phototube can easily distinguish between the control light and the background light.

When the mechanics of the installation require that the control light source be located at a consid-

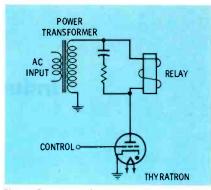


Fig. 3. Anti-chatter relay circuit.

erable distance, the phototube must "see" a weak control light against a strong background light. This problem is solved by the use of a control light whose characteristics differ from the background illumination. As shown in Fig. 5, the control light is modulated by a rotating disc. The circuit is so designed that it can distinguish this weak, flickering beam of light in the presence of a relatively constant background illumination. Systems of this type will operate reliably over distances of several thousand feet.

The motor-driven disc has a series of openings that "chop" the light beam at a rate determined by the number of holes and by the speed of the disc. Typically, the beam is chopped at a rate of 500 to 1500 hertz. Because the phototube amplifier is tuned to the chopping rate, it responds to the modulated light beam and ignores the steady background illumination. After amplification, the AC component of the phototube signal is rectified. The resulting DC potential biases the control tube below cutoff-or drives it into conduction if a light-actuated rather than a dark-actuated circuit is required.

The Shockover Capacitor

A shockover capacitor (Fig. 6) is connected between the grid and cathode of a thyratron. With the thyratron biased below cutoff, no current flows through the load until

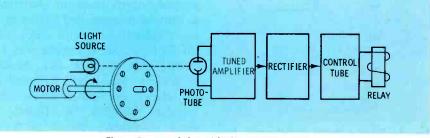


Fig. 5. Modulated-light photorelay.

46 PF REPORTER/January, 1967

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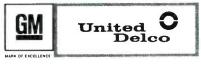
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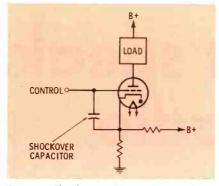
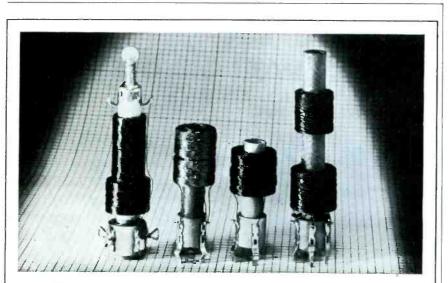


Fig. 6. Shockover capacitor used to prevent premature thyratron firing.

a positive control signal is applied to the grid. Were it not for the shockover capacitor, the thyratron would ionize as soon as the B+ voltage is applied (even if a positive control signal is not present at the grid). Such premature firing of the thyratron would occur because the thyratron behaves like a capacitive voltage divider. Since the cathodeto-grid capacitance and the grid-toplate capacitance are in series with the load and the B+ supply, the application of B+ voltage charges the resulting series capacitances within the tube. The charge that builds up across the cathode-togrid capacitance is positive at the



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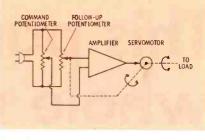


Fig. 7. Load position control employing servo system.

grid, and if high enough in amplitude, can fire the thyratron.

The voltage distribution in a capacitive voltage divider, is inversely related to capacitance; therefore, less voltage builds up across the larger capacitance. By increasing the cathode-to-grid capacitance of the thyratron, the shockover capacitor reduces the voltage to which this capacitance charges, and thereby prevents premature firing the thyratron. In addition, the shockover capacitor also serves to bypass spikes of noise voltage that might otherwise fire the thyratron.

Load Positioning Control

Servomechanisms are employed industrially for machine control and other load-positioning applications. As shown in Fig. 7, the mechanical load is positioned by a servomotor. This motor also drives the followup potentiometer through a speedreducing gear. The command potentiometer is the manual control which permits the operator to move the load to the desired position. Because of the mechanical feedback from the motor to the follow-up potentiometer, this potentiometer is automatically adjusted to match the setting of the command control. If these two potentiometers are not at the same setting, a difference of potential will exist between their sliders. This voltage is amplified to drive the servomotor. When the motor has adjusted the follow-up potentiometer to a setting corresponding to that of the command control, there is no longer a difference of potential between the two sliders. Since there is no input to the amplifier, the motor now stops.

The direction of rotation of the servomotor is determined by the phase of the driving signal from the amplifier; therefore, the motor rotates in a direction determined by the direction in which the operator has turned the command shaft. In this manner the operator can position a heavy load simply by setting the command control. For applications requiring automatic control, the command shaft is adjusted by a motion-producing device. For example, a bellows responding to changes of air pressure can supply the force to position a low-torque potentiometer.

Thermocouple Break Protection

Industrial temperature-controlers frequently employ thermocouples for sensing furnace temperature. The voltage from the thermocouple circuit is balanced against a reference voltage. When temperature is at the desired value, these two voltages cancel and no corrective action occurs. owever, if the thermocouple voltage is either larger or smaller than the reference voltage, the difference voltage represents the errors of the system. The difference voltage is use dto control a valve in the fuel line to the furnace, which adjusts the fuel flow to restore the temperature to the desired value. The difference voltage is also applied to a pen-type recording instrument which traces a graph of temperature versus time on a paper chart.

Because the thermocouple operates in a high-temperature environment, and in some applications is exposed to corrosive gases, it is subject to mechanical damage. If the thermocouple should open. It will no longer produce the voltage required to cancel the reference voltage. The circuit will interpret this as meaning "the furnace is too cold" and will cause the fuel valve to open wide. Consequently, the temp-



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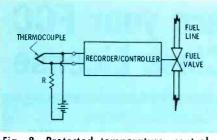


Fig. 8. Protected temperature control system using thermocouple.



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Circle 24 on literature card

erature of the furnace will rise above the desired value. Since the open thermocouple cannot sense this increase of temperature, the valve will open still wider and the temperature will increase still further. To safeguard the system against this runaway condition, a protective circuit is employed.

As shown in Fig. 8, a DC potential is applied to the thermocouple and series resistor R. Since the resistance of the thermocouple is extremely small compared to the value of R, practically all of the voltage appears across R. However, if the thermocouple should open, practically all of the voltage will appear across the open circuit. Because this voltage exceeds the reference voltage, the circuit interprets this to mean "the furnace is too hot". As a result, the circuit causes the fuel valve to close. In this manner, temperature runaway is prevented in the event the thermocouple opens.

Self-Latching Relay

Industrial circuits often require self-latching relay action. The relay energizes in response to a trigger signal and then remains energized after this signal is no longer present. This type of action can be accomplished with a mechanically locking relay. However, the same action can also be obtained with a relay of conventional construction as shown in Fig. 9.

A positive pulse applied to the grid drives the control tube into conduction. The relay now energizes and activates the controlled device—motor, lamp, bell, etc. An additional set of contacts establish another pathway for relay current through R and the reset switch. Since the relay coil now has a cur-

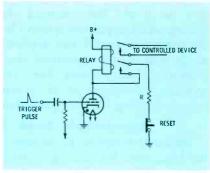


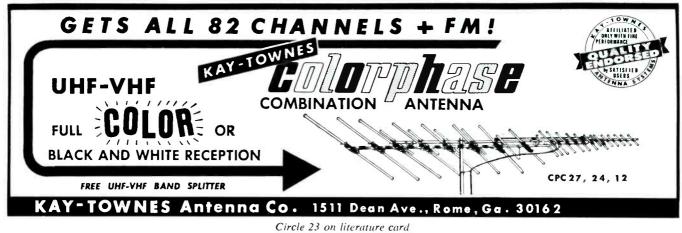
Fig. 9. Self-latching relay circuit.

rent path independent of the tube, the relay remains energized after completion of the trigger pulse to the grid.

By means of the reset switch, the operator can turn off the controlled device. The relay will now remain de-energized until another trigger pulse is applied to the tube. In some applications, the normallyclosed contacts of another relay are connected in place of (or in series with) the manual pushbutton. This second relay, when energized, will release the first relay.

Conclusion

The preceding paragraphs have pointed out the fact that industrial electronic circuits are not mysterious black boxes or nightmares of complex control circuits using exotic or unconventional electronic components. Although some of the circuits might be unfamiliar, the technician need not shy away from the field of industrial electronics as long as he has a thorough basic understanding of electronic components. Part II of this article will introduce other industrial applications of electronics, including the differential transformer, phase-controlled rectifier, saturable reactor, and conductivity meter.



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Bifilar Balun

(Continued from page 25) another eyelet, with nothing in it but one lead from the capacitor, and the center one at that. This must be the place, as the old burlesque joke says. Anyhow, what have I got to lose? So, I work the wires into this one, solder them, and crawl out from under the cabinet again.

It's not easy to change channels with crossed fingers, but I did. Now, low band; maybe a little better. Good. Now, (cross feet too) the high band. The high band booms in clear as a bell. It's hard to give a Tarzan yell and beat your chest without making a sound, but I did it. So, I checked them all out again, for luck-color, snow, and so on-and then called the widow. She came in with two cups of coffee and a plate of cookies (Told you she was my best customer, didn't I?) We sat down and watched the thing while I made out the bill (and that coffee cooled, She made the best coffee in town, and also the hottest!)

The moral of this story is: If at first you don't succeed, go back and check to see what kind of a mistake you made the first time. Seriously speaking, any of these balun coils can be rewound in an emergency. Because of their broadband nature, they are not as critical as coils in more sharply tuned circuits. As long as you get the connections, the number of turns, and the polarities of the windings right, they seem to do as well as the originals. The big headache, of course, is identifying the wires, particularly if your replacement is color-coded like mine was-all red. An ohmmeter and a couple of twists in the wire will help. Don't be afraid to recheck each connection to make sure that you do have the right wire in the right place. Above all, make up a rough sketch of the coil. core, and other parts before you take anything out. There will usually be enough insulation left around each wire so that you can tell "what went where" (exactduplicate replacements have the same color code as the original). However, rewinding can save you a lot of time, and make your "best customer" happy-this means a lot more than coffee and cookies.

Zip

Barber Pole (Continued from page 23) A. Plate

B. Cathode

Fig. 5. Burst pulse at phase detector.

tector include mismatched resistors R1-R2, a shorted or leaky C1-C2, or unmatched diode sections of the tube. (Some sets use solid-state diodes, which, of course, are also subject to mismatch).

Assume the reactance oscillator and phase detector stages are operating normally. They've passed our simple but conclusive tests. The out-of-sync color bars (barring the possibility of some "odd" problem) must be due to interruption or distortion of the incoming burst signal, in the burst amplifier stage.

Burst Amplifier Test

In our initial symptom analysis, we reasoned that some color burst was reaching the phase detector. Remember the bandpass stage is open, so the killer is at cut off; killer control voltage is initiated by the presence of color burst in the amplifier output. Our logical test then, is a waveform check for a clean burst signal of proper amplitude at the anode and cathode of the phase detector. For unbalanced operation of the phase detector with color sync applied, a similar burst signal must appear at both the anode and cathode of the detector. Figs 5A and 5B are typical of the burst waveforms at those points. Analyze the waveform checks as follows:

1. If the burst signal at either point is correct, the input circuit and operation of the burst amplifier tube can be considered proper. On the other hand, both signals missing or low in amplitude confirms improper operation of the amplifier.

2. If either waveform is missing or lower in amplitude than the other, the fault is a defect in the

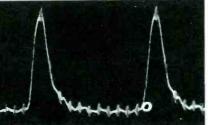


Fig. 6. Signal at grid of burst amp.

section associated with the wrong waveform. Possible causes include an open secondary section of T1, open C1-C2, or defects in the tint control network (and color killer coupling networks connected to the phase detector).

3. The burst transformer (T1) could be defective. A quick check can be made by performing the adjustment for T1 listed in the service data.

The burst amplifier functions much in the same manner as a sync separator. The tube is keyed on during horizontal retrace time; cut off during active scanning time. The grid receives both composite chroma video information and a keying pulse from the horizontal deflection system (Fig. 6).

If this waveform is low in amplitude, or the horizontal pulse is mistimed (color burst is not located at the upper tip of the keying pulse) the result could be a loss of color sync. However, the symptoms of this problem are usually good color lock on strong stations, poor lock on weaker stations.

Summary

Servicing for loss of color sync can be facilitated by following a systematic isolation approach. The approach isn't unfamiliar—it can be likened to the quick checks used when servicing horizontal oscillator and AFC stages.

See if the oscillator can freerun at 3.58 MHz by defeating the phase detector circuit. Check for a balanced phase detector output. Check for the presence of proper color burst (sync) from the output of the burst amplifier. If an odd problem arises, try performing the alignment procedure outlined in service information for the particular chassis. If any adjustment fails to perform its function, check the components directly associated with the adjustments.



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However, Brown may decide, because of the nature of his business, that the depreciation period be limited to five years, using a method that permits him to establish a larger depreciation reserve during the first five yea He would then use the "S Years-Digit Method" of depreciation, which uses ent fraction each year. add 1 + 2 + 3 + 4 + 4sum of the years would b tangible asset at the would have a useful li years, therefore the de for the first year would b

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on reserve	alternative, which wou	ıld permit
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He would	would show the following	
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be 5/15 of		3,600.00
	1967 Depreciation	
-	40% of 3,600	1,440.00
		2,160.00
	1968 Depreciation	

1968 Depreciation 40% of 2,160	864.00
1060 Dames	1,296.00
1969 Depreciation 40% of 1,296	518.40
1970 Depreciation	777.60
40% of 777.60	311.04
	466.56

cost, 4/15 for the second year, etc. The schedule would then appear as:

While Brown is evaluating meth-

ods to accumulate a large depre-

ciation reserve during the first years

of use, he might consider the "De-

clining Balance Method" as an

\$2.000

1,600

1,200

800

400

\$6.000

1966 Depreciation 5/15 of \$6,000

1967 Depreciation 4/15 of 6.000

1968 Depreciation

1969 Depreciation

1970 Depreciation

4.800

3/15 of 6,000

2/15 of 6,000

1/15 of 6,000

Conclusion

The foregoing discussion is intended, primarily, to acquaint the dealer or self-employed service technician with the methods of computing depreciation on test equipment, service vehicles, and other forms of tangible assets used in maintaining a business. The choice of a particular method will depend on such factors as the useful life of the equipment and the period of time over which the depreciation is to be spread. This, of course, can be determined only by the dealer or service technician.

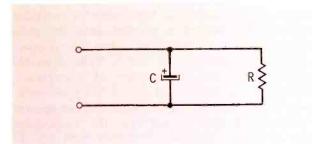


Fig. 14. Equivalent circuit of leaky capacitor.

series resistance present. Suppose the reactance is 100 ohms, and the series resistance is 60 ohms. Then, the impedance is 116.6 ohms. A capacitor checker indicates the *power factor* of the capacitor. The power factor is equal to the resistance in ohms (**R**) divided by the impedance in ohms (**Z**), or Pf = R.

A good capacitor has practically zero resistance. Therefore, the power factor is equal to O/Z = O. In other words, the power factor of a good capacitor is zero. On the other hand, suppose that a capacitor is nearly shorted; its impedance is practically the same as its resistance, and the power factor of the capacitor is equal to 1. We often speak of the power factor in percentage. Percentage means hundredths. For example, a power factor of 0.5 is equal to a power factor of 50%. Most capacitor checkers indicate the power factor as a percentage.

Equivalent Series and Parallel Resistance

When a capacitor is leaky, it is the same as if the capacitor were connected in parallel with a resistance, as shown in Fig. 14. This is a different configuration from that of Fig. 13. However, note carefully that every parallel RC circuit has an equivalent series RC circuit, as shown in Fig. 15.

- 1. The impedance of a series RC circuit is the hypotenuse of the triangle.
- 2. The impedance of a parallel circuit is the perpendicular dropped to the hypotenuse of the triangle.
- 3. If we have a given impedance Z, this impedance can be produced by R_s and X_s , or R_p and X_p .

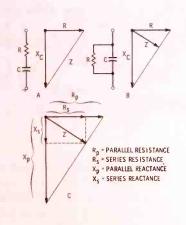
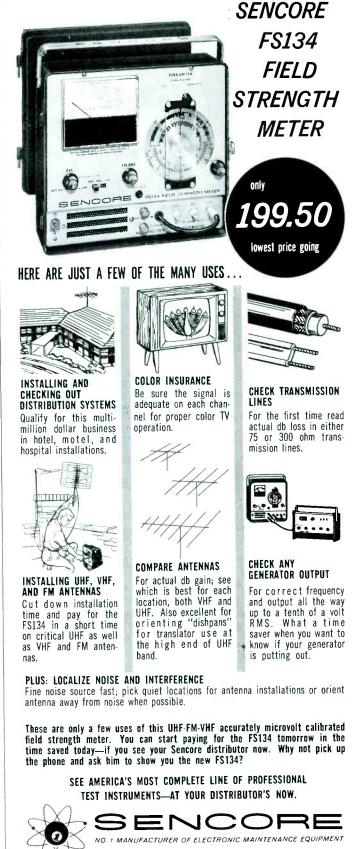


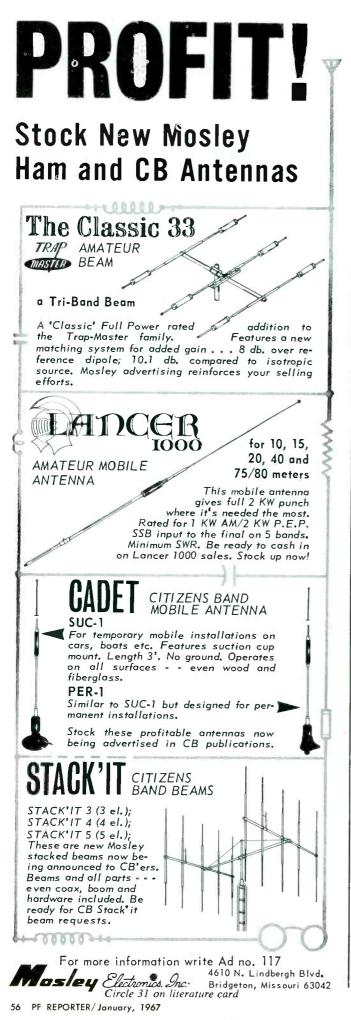
Fig. 15. Impedance of series and parallel RC circuits.

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January, 1967/PF REPORTER 55



We see that comparatively high values of resistance and reactance connected in parallel have the same impedance as comparatively low values of resistance and reactance connected in series. Your capacitor checker measures the power factor of a capacitor, whether the resistance is in parallel, in series, or whether some of the resistance may be in parallel and some of the resistance in series. In any case, the power-factor reading is correct. However, it follows from Fig. 15 that a power-factor measurement alone cannot tell us whether the capacitor is leaky (parallel resistance present). Therefore, your capacitor checker also has a leakage test, to determine whether a poor power factor is being caused by leakage.

Conclusion

We have seen how Ohm's law applies to AC circuits, and how AC voltages are added together. The meaning of reactance and impedance is evident, and we know how to find the reactance of a capacitor or an inductor by using a simple chart. We also recognize the meaning of power factor, and why a poor power factor does not necessarily mean that a capacitor is leaky. In the future, we will dig a little deeper into Ohm's law for AC circuits, and explain measurements of resistance, reactance, and impedance at broadcast, HF, and VHF frequencies. Such explanations should greatly simplify a number of practical problems confronted at the service bench.



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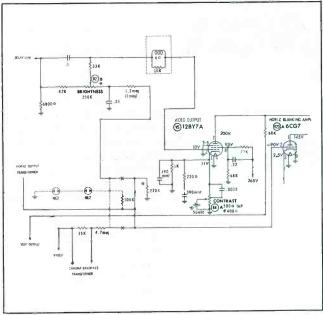
Shadows and Increased Contrast

I am having double trouble with an RCA CTC11D chassis (PHOTOFACT Folder 550-2). What appear to be shadows flitter across the screen, particularly during warmup. The shadows do not pull or bend the raster or picture, but seem to be superimposed upon the picture.

The other trouble symptom is intermittent brightness modulation, with the picture becoming dark. Both trouble symptoms occur during both color and black-and-white reception. All tubes, including the picture tube, check good. I have been unable to detect any discrepancies in voltages or waveforms. I have recently replaced two multiple section filters in the chassis. Do you have any ideas on the possible cause or causes of the foregoing symptoms.

Cupertina, Cal.

WALTER W. ELLITT



Your description of the shadow symptoms has not made it clear whether the shadow moves across the screen vertically or horizontally. A horizontal shadow or bar would indicate a slight 60- or 120-hertz hum caused by insufficient filtering in the low-voltage power supply. However, such hum normally produces a certain amount of picture bending or slight tearing as it moves across the screen. Since you have recently replaced what I assume to be the multiple section electrolytics in the low-voltage supply, it is possible that a had solder connection is reducing the filtering action of one of the electrolytic sections. A quick scope check of the six low-voltage sources should provide the answer. If the shadow moves vertically across the raster, it could be caused by some form of external interference (a mild windshield wiper effect etc.). There are also internal causes of interference which could cause the same



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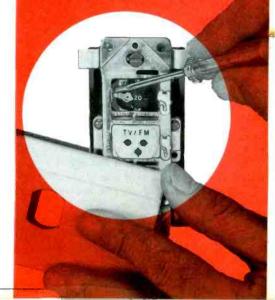
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symptom, such as video getting into the horizontal sweep circuit.

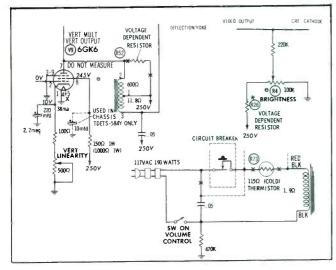
The darkening of the picture sounds like excessive contrast caused by either a video overload or a defect within the brightness or contrast circuits. While it is doubtful that an AGC circuit defect is causing the trouble, it would not hurt to eliminate this section as a possible source of trouble by clamping. The most likely source of trouble will be found in the unique brightness control circuit (shown here) employed in this chassis. Early versions of this chassis used two NE2 neon lamps to provide regulation of the fixed hias supplied to the grid circuit of the video output tube and chroma bandpass amplifier. The many instances of failure of these neon regulators has produced insufficient and intermittent brightness symptoms, which at the same time seemed to be caused by the video circuit. If the problem chassis uses the neon lamps, the circuit should be modified as shown on the schematic. Another possible cause of what appears to be excessive contrast or overload is a defective 2-mfd capacitor from the cathode to screen of the first video amplifier. This overload effect is particularly noticeable when the set first comes on. A quick check can be performed by clipping one end of the capacitor-the set operates better without the capacitor than it does with a defective one.

Delayed Height

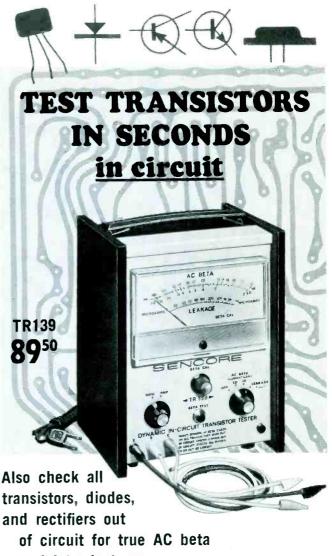
The raster of a Motorola Chassis WKTS-584 (PHOTO-FACT Folder 692-2) fails to expand to full height for 5 to 8 minutes after the set is turned on. What are the most likely faults that will cause slow warmup symptoms.

Cincinnati, Ohio

ERNEST BONEM



Defective thermistors and voltage dependent resistors (VDR) are possible causes of slow warmup. The trouble symptom will usually pin point the defective component. As an example: In the case of delayed height, such as you describe, the most probable fault exists within the vertical sweep circuit (R52, the VDR in the plate circuit of the vertical output stage). If the slow warmup symptom was a dim raster, the most likely candidate would be R26, the VDR in the brightness control circuit. A combination of symptoms such as reduced height, unstable sync, and dim raster would probably be caused by a defective component in the low-voltage power supply (thermistor R73). There are other causes of slow warmup, such as increased resistance of a fixed heater-voltage dropping resistor, or a marginal cathode bypass capacitor or input filter capacitor.



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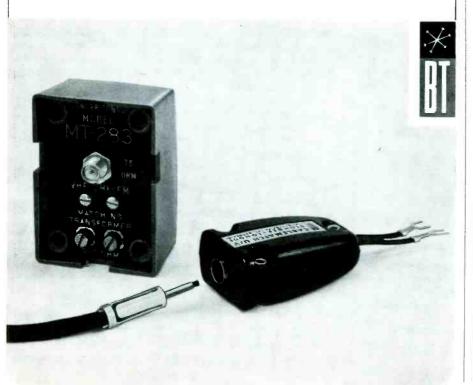
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Quality matching transformers like these are just one more reason why you should go Blonder-Tongue from antenna to TV set terminals. Write for free catalog #74.

Blonder-Tongue Laboratories, Inc., 9 Alling Street, Newark, N.J. Blonder-Tongue-the name to remember, for TV reception you'll never forget

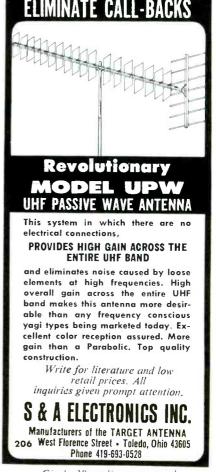


62 PF REPORTER/ January, 196 Fircle 39 on literature card



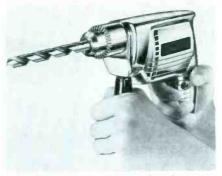
Tuner Cleaner

A new tuner cleaner, formulated to cope with the added sensitivity of nuvistor and transistor tuners, is announced by **Chemtronics**. Lab tests of the new No. 800 Tun-O-Lube Nuvistor/Transistor Tuner Cleaner were performed by injecting a simulated



Circle 38 on literature card

TV signal into TV tuners, using a 6MHz bandwith. The resultant waveform was observed on a scope while the tuner cleaner was applied over and over. No measurable frequency deviation was noted on the scope. Lab tests have been going on for over 18 months on all new TV tuners as fast as they are introduced on the market. Simultaneously with lab tests, key servicing centers handling large numbers of sets of all makes were chosen and similar tests performed on actual 'in-repair" sets. The tuner cleaner is safe to use on all plastics used in TV tuners. Price is \$1.98.



and ball thrust spindle bearings assure smoother running performance. The motor and other power controlling components are contained in a compact, die cast aluminum housing with life-time, tarnish-proof Ball-Brite finish.

The unit is equipped with a $\frac{1}{2}''$ geared Jacobs chuck with key, locking trigger with safety release, and 6 ft three-wire cord. It weighs $5\frac{1}{4}$ lb. and measures $9\frac{1}{4}'' \ge 7\frac{1}{4}'' \ge 2\frac{1}{2}''$. The drill comes packaged in its own plastic encased container with a four-color labeled outer shipping carton showing the tool in action and applications. Price of the $\frac{1}{2}''$ drill is \$29.95.



Compact Power Drill

A new compact, $\frac{1}{2}$ " drill with a 3.5-ampere, 115-volt AC high-torque motor which operates at a constant speed of 1100 RPM, has been introduced by **Wen Products, Inc.**

The high-torque motor is designed with welded burnout-proof armature. Needle bearings at heavy thrust points

Image: A mage: A mage:

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NUTLEY, NEW JERSEY 07110

Circle 41 on literature card

Perk it up with Perma-Power COLOR-BRITE

Perma-Power does for color TV sets what we've done for millions of black and white CRT's: adds an extra year of useful picture tube life.

When a color tube begins to fade, COLOR-BRITE *instantly* brings back the lost sharpness and detail. It provides increased filament voltage to boost the electron emission and return full contrast and color quality to the 3 gun color picture tube.

COLOR-BRITE is automatic . . . no switching or wiring. Just plug it in. Your delighted customers will brighten up as fast as their color sets!

Model C-501, for round color tubes. List Price \$9.75 Model C-511, for rectangular color tubes. List Price \$9.75



5740 N. TRIPP AVE., CHICAGO, ILLINOIS 60646 PHONE (312) 539-7171 Circle 40 on literature card

January, 1967/PF REPORTER 63



COLOR-BRITE is a Hue-Brite product from Perma-Power. famous in TV service for b & w Vu-Brites and Tu-Brites. Nowhere Else Can You Get All Of These Features For Less Than \$875



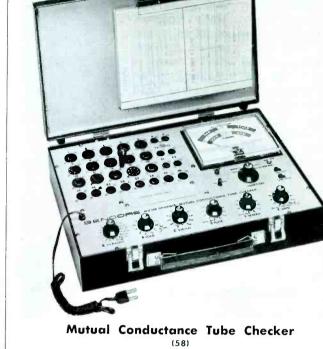
The New Heath IO-14... a 5" DC to 8 MHz Scope with continuous duty ratings, triggered sweep, 0.25 u sec coaxial vertical input and time base ...factory assembled & tested for \$399.00 ... kit only \$299.00

The Heath IO-14 Features The Engineering And Quality Components You Expect Only In Higher Priced Oscilloscopes. For example, switches are ball-detent type; all major control potentiometers are precision high-quality sealed components; all critical resistors are 1% precision; and vertical signal delay is provided through highlinearity coaxial delay lines. Here is the ultra-stable, low-noise performance demanded from a truly professional industrial, academic, electronic engineering laboratory oscilloscope. Check the prices and specifications yourself, and you'll agree the Heath IO-14 gives a new meaning to 'scope value.

Kit IO-14 (with standard P-2 phosphor), 53 lbs
Assembled IOW-14 (P-2 phosphor), 47 lbs\$399.00
Assembled IOW-14S (with long persistance P-7 phosphor for bio-medical or industrial use), 47 lbs
Kit PK-1, Low-Capacitance Probe, 1 lb \$4.95

10-14 SPECIFICATIONS - (Vertical) Sensitivity: 0.05 v/cm AC or DC. Frequency response: DC to 5 MHz, -1 db or less; DC to 8 MHz, -3 db or less. Rise time: 40 nsec (0.04 microsecands) ar less. Input impedance: 1 megohm shunted by 15 uuf. Signal delay: 0.25 microsecond. Attenuotor: 9-position, compensated, calibrated in 1, 2, 5 sequence from 0.05 v/cm. Accuracy: $\pm 3\%$ on each step with continuously variable control (uncalibrated) between each step. Maximum input voltage: 600 volts peak-to-peak; 120 volts provides full 6 cm pattern in least sensitive position. (Horizontal) Time base: Triggered with 18 calibrated rates in 1, 2, 5 sequence from 0.5 sec/cm to 1 microsecond/cm with ±3% accuracy or continuously variable control position (uncalibrated). Sweep magnifier: X5, so that fastest sweep rate becomes 0.2 microseconds/cm with magnifier on. (Overall time base occuracy $\pm 5\%$ when mognifier is on.) Triggering capability: Internal, external, or line signals may be switch selected. Switch selection of + or - slope. Variable control on slope level. Either AC or DC coupling. "Auto" position. Triggering requirements: Internol; ½ cm to 6 cm display. External: 0.5 volts to 120 volts peak-to-peak. Horizontal input: 1.0 v/cm sensitivity (uncolibrated) continuous goin control. Bondwidth: DC to 200 kHz ±3 db. General: 5ADP81 or 5ADP2 Flat Face C.R.T. interchangeable with any 5AD or 5AB series tube for different phosphor chorocteristics. 4250 V. occelerating potential. 6 x 10 cm edge lighted graticule with 1 cm major divisions & 2 mm minor divisions. Power supply: All voltages electranically regulated over range of 105-125 VAC or 210-250 VAC 50/60 cycle input. (Z Axis) Input provided. DC coupled CRT unblanking for complete retrace suppression. Power requirements: 285 watts. 115 or 230 VAC 50-60 Hz. Cabinet dimensions: 15" H x 101/2" W x 22" D includes clearance for hondle and knobs. Net weight: 40 lbs.

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And a second	FREE	Benton Harbor, Michigan 49022
HEATHKIT 1967	1967	Please send FREE Heathkit Catalog & Information describing the New Heathkit
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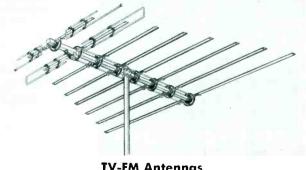
A new mutual conductance tube checker, using a 5-kHz squarewave for Gm tests, has been announced by **Sencore**. Compact and completely portable, the all-purpose MU140 Continental is designed for fast, accurate, true mutual conductance tests with readings in actual micromhos.

Featuring an automatic biasing system, the Gm tester circuit actually biases the tube with the correct bias voltage at the current selected for the plate circuit. This eliminates a set-up control, and reduces possible errors in set-up and readings.

In checking amplifiers and rectifiers, the unit pulls the actual rated cathode current from the tube under test to measure its current emitting capabilities. The tube tester also provides hi-sensitivity grid leakage tests to spot leakage as small as 100 megohms, plus shorts tests that reveal shorts of 180k ohms between elements.

The unit can be used for checking all TV and radio tubes, including Novars, Compactrons, Nuvistors, Magnovals and foreign tubes. Space is also provided for additional sockets to accommodate future tubes with different base arrangements. Protection against obsolescence is thus assured.

The attache-type case is made of mar-resistant vinylclad steel. The unit, complete with tube chart, is priced at \$179.50.



TV-FM Antennas

A new line of television antennas is now available from Jerrold Electronics. Called Paralog-Plus, the new series of

antennas utilize a Bi Modal director system in which the parasitic elements combine two high-band directors into a single director covering the entire low band, plus all FM channels. The driven elements work in two modes simultaneously: $\frac{1}{2}$ wavelength mode for low-band channels and 3/2 wavelength mode for high-band channels. Thus, each element serves double duty.

Characteristics of the new antennas include a response of ± 1 dB per channel and impedance matching to both 300- and 75-ohm outputs. Mechanical features include self-cleaning wedge snap locks, which actually tighten in wind; Cycolac insulators, providing 4" separation of feed lines to eliminate shorting due to salt spray or ice build-up; a corrosion resistant finish; dual square boom construction; and grounded transmission lines to prevent build-up of static electricity.

The eight models in the new series range in price from \$17.95 for Model PIX-35 with 4 driven elements, to \$81.95 for Model PIX-225 with 10 driven elements and 9 parasitic elements.



Color TV Service Aid

A new service aid which permits the serviceman to observe an enlarged view of the TV screen while working on controls in the back of the set is introduced by the B&K Division of the Dynascan Corp. Dyanascope consists of a 21/4" diameter mirror attached to one end of a 6" stainless steel rod with a suction cup at the other end. Mounted in the middle of the rod is a $1\frac{1}{2}$ " diameter, three-power, magnifying lens. The serviceman simply attaches the unit to the TV screen with the suction cup and adjusts the mirror and lens for viewing from the back of the set. Then, while adjusting controls for static convergence, dynamic convergence, and color purity, he looks over the top of the set to observe the changes on the screen. The wasted time and motion of having to constantly walk back and forth between the front screen and rear controls is eliminated.

The unit is furnished with a plastic carrying case, designed to fit easily in a tube caddy or other service case. To simplify storage and prevent damage, the mirror folds down and the magnifying lens can be removed from the rod for insertion into a special protective pocket in the carrying case. Price of the unit is \$7.95.

New! All-channel COLORITE 300 ohm shielded

cable

CAT. NO. 847

Assures superior color reception on all 82 channels.
Eliminates costly matching of transformers and coaxial cable fittings.

• Reduces radiation and outside interference.

• Eliminates transmission line pick-up of ghost signals and automobile noise.

• Attached drain wire. Easy to install anywhere.

Low loss twin cellular polyethylene tubes encapsulated in aluminum foil shield. Spaced for proper impedance. Non-contaminating, non-deteriorating brown polyethylene jacket. Get the facts on this all-new, all-channel cable now. Send for FREE SAMPLE and spec. sheet.



Circle 42 on literature card January, 1967/PF REPORTER 65 "I was TROUBLE-SHOOTING in the DARK 'til Sencore's Scope Showed Me What Color Waveforms Really Look Like."



Technicians everywhere are talking about the PS127 5" Wide Band Oscilloscope. Try one and you, too, will send us comments like these-

"So easy to use! With my Sencore scope I can read high or low frequency signals without band switching. As easy to use as a voltmeter."—R. L., Portland, Ore.

"I've only had my PS127 a couple of months, but it's more than paid for itself already with the extra jobs I've been able to handle." -S. O., New Orleans, La.

"With the direct peak-to-peak readout I can compare voltage readings to those on the schematic without wasting valuable time setting up my scope with comparison voltages." - J. M. F., Plymouth, Michigan.

"Those Sencore exclusives really sold me, like the extra 500KC Horizontal Sweep range and the free high voltage probe."-D. N., Brooklyn, N.Y.

"With the PS127, I find I can trouble-shoot those tough ones twice as fast as before—especially color TV."—F. C., Burlingame, Calif.

"Once I compared the specs, I knew Sencore had the best buy in scopes. We now have three PS127's in our shop."—J. S., Ft. Lauderdale, Fla.

SPECIFICATIONS



Inverter (61)

This new inverter which plugs into a cigarette lighter socket, is specially designed to compensate for lowpower factor in tape recording. The transistorized unit can be used to operate TV sets, lights, tools and many household appliances. It can also be used for citizen's band equip-

ment and public address systems. The Electro inverter delivers 125 watts continuous AC power from a 12-volt car, boat, or plane battery, or from any 12-volt DC source.

Features of the unit include: automatic overload protection, a charge indicator light that shows the condition of the battery, and gives a lowcharge warning, and a "start" switch that provides rapid starting of light bulbs, motors, etc. A battery clip adapter kit is available for connecting directly to battery terminals if use of the lighter socket is not desired. Units are supplied with complete operating instructions, and performance curves. The unit measures 31/2" $x 6\frac{1}{4}$ " $x 6\frac{1}{4}$ " and weighs $6\frac{3}{4}$ lbs. Price is \$54.50.



Circle 43 on literature card



Event Counter

The new DP-140 Event Counter and Slave Plug-in extends the application of Hickok's DMS-3200 Digital Measuring System to provide inexpensive, all-electronic counting and display of totalized count of events in both industrial and laboratory applications. A single plug-in/main frame combination provides a three-digit display and additional combinations may be added in cascade to provide six-, nine-, or twelve-digit display. The combination may also be used to extend the readout of Hickok's DP-150# 1-MHz Frequency Counter to a simultaneous six-digit display of frequency or period. Operation may be remotely controlled to "start count," "stop count," "resume count," or "reset." Sensitivity is 100 millivolts.

The unit is small in size and weighs only 2³/₄ pounds. Price of the DP-140 is \$75.00. Price of the DMS-3200 Main Frame is \$320.00.



Miniature General-Purpose Probe (63)

A miniature, 10X attenuation probe, designed for portable and general-purpose oscilloscopes in the DC to 33 MHz range, has been announced by **Tektronix**, Inc. The P6012 has a convenient adjustment that enables it to compensate for oscilloscope or plug-in unit input capacitances of 15 to 47pf. The small size and lightness of the probe makes it particularly useful in applications where accessibility to circuitry is difficult.

Risetime of the probe is approximately 5 nanoseconds. Input resistance is 10 megohms. Input capacitance is approximately 11.5pf, with a 3.5-foot cable and approximately 14.5pf with a 6-foot cable. Voltage rating is 500 V DC, AC peak, or DC and AC peak combined. Peak-voltage derating is necessary for CW frequencies above 6 MHz — typically 105 V at 30 MHz. Price of the probe is \$27.00.

CORRECTION!

The Triplett Model 630 advertised on page 75 in the November PF RE-PORTER should have been priced at \$55, and the Model 630A at \$65.

Now, for men in electronics –"a whole new era of quick calculation"



THERE MUST BE THOUSANDS OF PEOPLE in electronics who have never had the marvelous adventure of calculating problems with a single slide rule; other thousands have had to content themselves with a siide rule not specifically designed for electronics. For both groups, the new slide rule designed and marketed by Cleveland Institute of Electronics and built for them by Pickett will open a whole new era of quick calculations.

"Even if you have never had a slide rule in your hands before, the four-lesson instruction course that is included takes you by the hand and leads you from simple calculations right through resonance and reactance problems with hardly a hitch. If you already use a slide rule, you'll find the lessons a first-rate refresher course. And it explains in detail the shortcuts built into this new rule."

From an article in Radio Electronics Magazine

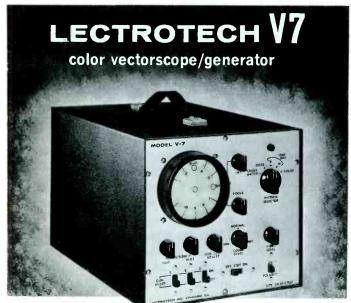
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Circle 46 on literature card

Why is a Vectorscope essential for Color TV servicing?

- Check and align demodulators to any angle . . . 90°, 105°, 115°... accurately and quickly. No guesswork. New color sets no longer demodulate at 90°. Only with a Vectorscope can these odd angles be determined for those hard-to-get skin tones.
- 2 Check and align bandpass-amplifier circuits. Eliminate weak color and smeared color with proper alignment. No other equipment required. Only a V7 Vectorscope does this.
- Pinpoint troubles to a specific color circuit. Each stage in a TV set contributes a definite characteristic to the vector pattern. An improper vector pattern localizes the trouble to the particular circuit affecting either vector amplitude, vector angle or vector shape. Only a V7 Vectorscope does this.

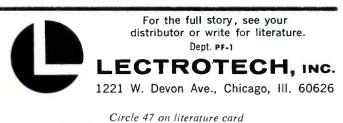


EXCLUSIVE FEATURES:

Color Vectorscope: Until now, available only in \$1500 testers designed for broadcast use. Accurately measures color demodulation to check R-Y and B-Y, for color phase and amplitude. A must for total color and those hard-to-get skin tones. Self-Calibrating. Adjust timing circuit without external test equipment. Dial-A-Line. Adjust horizontal line to any width from 1-4 Jines. Solid State Reliability in timer and signal circuits. Plus: All Crosshatch, Dots, Vertical only, Horizontal only and Keyed Rainbow Patterns. RF at channels 3, 4 or 5. Video Output (Pos. and Neg. adjustable) for signal injection trouble-shooting. Red-Blue-Green Gun. Killer. All transistor and timer circuits are voltage-regulated to operate under wide line voltage ranges. Lightweight, compact—only $8\frac{1}{4}x7\frac{1}{2}x12\frac{1}{2}$ ". NET **189**⁵⁰

ONE YEAR WARRANTY

V6-B New, improved complete color bar generator with all the features of the V7 except the Vectorscope. Only 99.50



Color Countermeasures

Symptoms & Tips From Actual Shop Experiences.

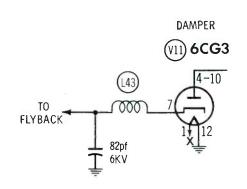
Chassis: ZENITH 26KC20

- **Symptoms:** Diathermy type RF interference in picture when tuned to higher channels and radiating into other sets in the near vicinity. Disabling the high voltage eliminates the trouble.
- **Tip:** Dress and reroute wire on pin #6 of 6JH8 B-Y demodulator tube if necessary. If not already installed, insert a ferrite bead, Zenith part #149-171 on the wire where it junctions at the tube socket pin.

Chassis: GE CB (early runs)

Symptom: Insufficient width

Tip: Add an 82 pf 6KV cap from junction of L43 and flyback to ground. This cap has been added to all late production runs.



Chassis: Motorola TS 914

www.americanradiohistory.com

Symptom: Weak color, color indicator light goes out when color control is turned up. Tip: Check for open L21.

IST CHROMA BANDPASS AMP (1) A 6DX8/ECLB4 (1) B 6DX8/ECLB4 (1) B

68 PF REPORTER/January, 1967

Notes On Test Equipment

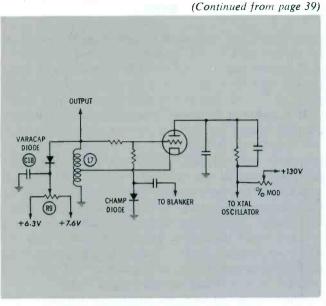


Fig. 4. Simplified oscillator schematic.

sures that the marker amplitude will not be too great if you happen to get a super-active crystal. The audio cable has ben eliminated, and instead there is now an RCA-type phono jack at the front panel. This jack is used for both audio out and external modulation in.

The biggest news in the "B" through, is the addition of sweep outputs. The B model costs about \$5 more than the A did, and for this, the purchaser gets a sweep generator in addition to the RF generator.

The oscillator is shown in Fig. 4. This is a much simplified diagram, as the oscillator does double duty, both sweeping and as a conventional Hartley oscillator, an dcovers eight separate bands. The heart of the sweeper is the tank circuit made up of L7, C18, and CR3. At rest, this circuit resonates at 455 kHz, as CR3 acts as a small-value capacitor. C18 is a relatively large value capacitor and is essentially a short circuit for RF, thereby grounding the anode of CR3. An AC voltage applied to CR3 causes it's capacity to vary, proportional to the voltage change. R9 is a sweep linearity control; the DC voltage applied to CR3 through R9 tends to distort the capacity curve change, according to the position of the control arm.

The change of capacity of CR3 changes the resonant frequency of the tank and therefore the output frequency of the oscillator. A blanking voltage is applied to the grid of the oscillator. Through careful choice of components, the oscillator is allowed to conduct only while the frequency is rising. As the voltage across CR3 reverses and the output frequency starts to decrease, the tube is cutoff and does not resume oscillating until the tank frequency again begins to rise; the net result is a scope trace showing a zero base line.

The sweep frequency and blanking are keyed to line frequency, therefore the unit can be used with any scope which has a line sweep position. (Providing it doesn't present too much phase shift.)

The sweep width is about 10% of center frequency, so the bands swept are about 400-500 kHz, and 10.2-11.2 MHz. Any crystal in these bands may be used as a marker by merely plugging it into the Xtal jack.

You can pay more but you buy less!

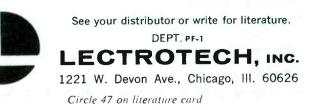


new Lectrotech V6-B color bar generator

9950 brings you...

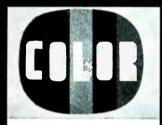
- Guaranteed performance . . . full one year warranty.
- A.C. operated solid state generator . . . no batteries to buy . . . no failures when needed most.
- Most stable generator of all . . . sharper, brighter patterns.

The V6-B provides crystal-controlled keyed rainbow color display, all cross hatch, dots, vertical lines only, horizontal lines only, Red-Blue-Green gun killer, exclusive Dial-A-Line feature (horizontal line width adjustable), voltage-regulated transistor and timer circuits, simplified rapid calibration. Supplies adjustable dot size, RF output more than 10,000 mv, operates on channels 3, 4 or 5, has color level control for color sync servicing. Connects to antenna terminals (no connection needed inside of set). Power transformer-line isolated. Stable operation under wide voltage ranges assured by fully voltage-regulated circuits. Hand-wired reliability . . . no printed circuits. Fully enclosed test lead compartment. Rugged, caddy size unit built to withstand rigors of field servicing. Size: 75_{6} "W, 31_{2} "H, 9"D. **9950**



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Scanner

(Continued from page 16)

Sales

A 41% sales gain for the first nine months of 1966 was reported by **Amphenol Corporation.** Net earnings for the period were \$4,224,610, compared to \$2,537,-427, for the same period of 1965. Consolidated sales for the first three quarters were \$113,146,408 compared to \$79,995,885 for the first nine months of 1965.

E. F. Johnson Company announced record sales and earnings figures for the nine months ended September 30. Sales were \$10,851,998, an increase of 18.7% compared with \$9,139,349 for the same period last year, and after tax net earnings rose to \$1,380,902, an increase of 26.3% over last year. In commenting on the company's continuing favorable growth rate, President Edgar F. Johnson says, "We have no worries about a 'peace depression' because of a very small reliance on military orders. The radio communications field is acknowledged to be expanding and Johnson's position in it is strong and growing soundly."

Nine months' net sales and earnings of Littelfuse, Inc. were at record levels. Sales were \$8,667,148 compared with \$7,202,119 for the like period in 1965. Earnings were \$268,558 whereas in 1965 they were \$261,329.

Third quarter and nine month sales and earnings of **Oak Electro/Netics** reached record levels, with third quarter net income tripled over the 1965 period. Consolidated net sales for the first nine months of 1966 to-taled \$47,122,018, a 20% increase over the \$39,107,-043 reported last year. Nine months earnings increased 80% to \$1,578,864, compared with the \$875,479 in the same 1965 period.



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INDEX TO ADVERTISERS

JANUARY, 1967

Amphenol Corp. 31
B & K Mfg. Co., Division of
Dynascan Corp
Blonder-Tongue Labs, Inc
Bussmann Mfg
Castle TV Tuner Service
Chemtronics, Inc
Clarostat Mfg. Co., Inc. 10
Cleveland Institute of Electronics 15, 67
Delco Radio
Electro-Voice, Inc. 11
The Finney Co
Grantham School of Electronics 54
Heath Co. 64
Hickok Electrical Instrument Co.
International Wire and Cable
JFD Electronics Co 40
J. W. Miller Co
Jerrold Electronics Corp cover 2
Kay-Townes Antenna Co 50
Lectrotech, Inc

Littelfuse, Inccover	4
Mosley Electronics Inc.	
National Radio Institute	
Oaktron Industries, Inc.	37
Oxford Transducer Co.	.53
Perma-Power Co.	63
Planet Sales Corp.	63
Quam-Nichols Co.	71
Quietrole Co.	66

RCA Components and Devices

(Entertainment Receiving Tubes)

cover	3
RCA Institutes, Inc	0
RCA Sales Corp	5
S and A Electronics	2
Sams, Howard W. & Co., Inc52, 7	1
SENCORE, Inc. 27, 34, 55, 57, 59, 61, 60	6
Sprague Products Co	4
Sylvania Electric Products Co	0
Texas Crystals	0
Weller Electric Corp	2
Winegard Co	2
Zenith Radio Corp 10	6



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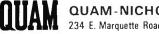
MODEL 8C10PA

For top quality public address. This 8" speaker has a 10 oz. ceramic magnet, frequency re-sponse of 55-13000 Hz., easily handles 12 watts. One of 34 Quam speakers for sound system applications.

Men who specify loudspeakers in enormous volume have to be particular about quality and performance. That's why more and more sound system installers are developing the habit of specifying Quam.

It's a good habit to acquire, because Quam makes good speakers for public address, background music and other sound system needs, as well as for radio-tv-automotive replacements.

Whatever kind of speaker you need, look for Quam, the Quality line, in the red, white, and blue package at your distributor.



QUAM-NICHOLS COMPANY 234 E. Marquette Road-Chicago, Illinois 60637





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- 127
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- 129.
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- 132.
- SIMPSON—Flyer giving specifications of Model 604 multicorder for measuring and recording volts, amps, milliamps, and microamps. 134
- TRIPLETT-Catalog D-66-I features the complete line of panel instruments. 135.

TOOLS

- ARROW-Literature describing 3 staple 136. guns.
- ENTERPRISE DEVELOPMENT—Time-saving techniques in brochure from En-deco demonstrate improved desoldering and simplifying operations on PC boards. 137
- 1.38. 139
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