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Construction Project

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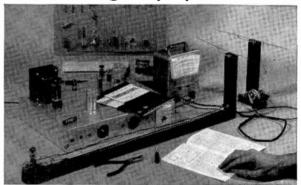


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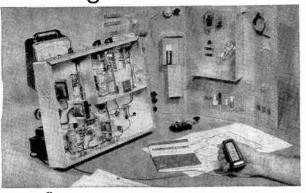
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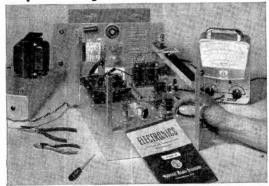
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This month's cover photo by Conrad Studio

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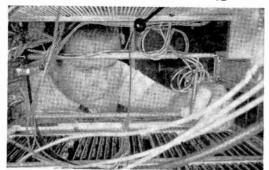
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CIRCLE NO. 28 ON READER SERVICE PAGE

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COMMUNICATIONS SATELLITE CORPORA-TION – the unique commercial communications company authorized by Congress to establish a global communications satellite system – has immediate needs for electronic technicians to work in its rapidly expanding laboratories.

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CIRCLE NO. 12 ON READER SERVICE PAGE

letters

FROM OUR READERS

ANGRY VOICE ANGERS READERS

I am surprised that POPULAR ELECTRONICS would publish "Angry Voice from Red Lion" (September 1969, p 73) which suggested that short-wave broadcaster WINB be taken off the air. It displays the usual intolerance of liberals who despise anything critical of communism, however true.

H. E. Foss Coulee Dam, Wash.

I did not like the article and I'm glad my subscription ran out.

CARL MILLER Greensboro, N.C.

The slander of "Lifeline" and "The Manion Forum," which are simple and relatively pure expressions of traditional Americanism, clearly indicates the evil motivations of the author.

V. L. CHAPPELL Los Angeles, Calif.

I'm sure that there are many anti-communist listeners in foreign countries who are not at all confused, but greatly encouraged by the true American culture reflected over WINB.

> MARGARET E. LANGRIDGE Geneseo, Ill.

POPULAR ELECTRONICS is not a magazine for airing personal political opinions.

Y. SAWERDA San Diego, Cal.



At least Dr. Carl McIntyre isn't trying to get the World Council of Churches, or others, thrown off the air, because he believes in freedom with responsibility, not irresponsible freedom.

K. VEVERKE Central City. Iowa

The story is an excellent communist promotion of the limitation of freedom by right-wing conservatives.

Name Withheld

I would welcome the opportunity of WINB to respond to your editorial.

Name Withheld

Reader comments regarding the "Red Lion" story were numerous and easily placed in two categories: Popular Electronics should be a non-political magazine; and the story was a communist "plant" Neither opinion descries editorial recognition or rebuttal. No letter has been received supporting the FCC viewpoint that radio programs for "domestic" consumption should not be aired on shortwares.

CB! YUK!

I monitored CB from a hill near the Bernhiem Forest in Kentucky and every station heard was outside of Kentucky with the closest being in Nashville—about 150 miles distant. One station in Fairfield, Alabama was "in contact" with a fellow in Findlay, Ohio. Fourteen more were in a roundtable on one channel and at this distance none of them could have been operating with legal power. None signed legitimate callsigns, but used fictitious identification such as "Grand Ole Pappy," "Mississippi John Henry," etc. Obviously, CB'ers do as they please—the

Obviously, CB'ers do as they please—the FCC notwithstanding. I say let's close CB down or make every licensee show a useful cause for his having a CB rig.

W.R. YEARY, K4ABL/K5HZG Louisville, Ky.

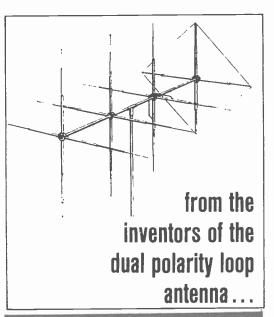
There is no argument from Popular Electronics that a minor percentage of CB'ers is making this whole communication medium look ridiculous. Obviously, there are many healthy and helpful aspects, but the legitimate CB'ers cannot clean house without the support of the FCC—which appears to let "things slide."

ENCAPSULATION

I have been using the plastic resin casting technique ("Encapsulate Your Circuit," August 1969, p 82) for a year. But be careful of hardening the casting with a hair dryer, I found that encapsulating the "PPFL" project, shown in the same issue, created some problems when the resistors upped their values alarmingly.

JACK KOLTON Reading, Pa.

(Continued on page 10)



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Power Handling:	1000 watts
Front to Back	
Separation:	
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*Patent Pending	Made in America





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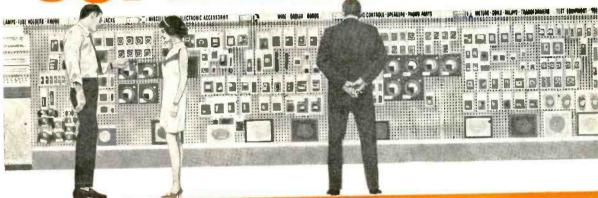
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Attractively packaged 2 per pack, showing

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CIRCLE NO. 16 ON READER SERVICE PAGE

LETTERS

(Continued from page 9)

WHY EDISON ROULETTE AT ALL?

I commend you and the author of "Why Play Edison Roulette?" (August 1969, p 71) for one more warning that every home can be a death house.

However, aren't we treating the wrong thing? Shouldn't we look at the real criminal



-the improperly manufactured electrical appliance?

The public should be told that the conventional twin-wire power cord is a do-it-your-self death tool. The public must be made aware that a three-wire power cable is the only route to life assurance.

> C.O. TAYLOR San Diego, Cal.

SERVICE FOR THE BLIND

Our contact with the blind community indicates to us that many blind persons are unaware of the fact that POPULAR ELECTRONICS is available to them on tape from Science for the Blind. This is true in spite of the fact that we have tried to notify the community through notices in braille periodicals and through direct mail to our own mailing list.

Perhaps you and your readers can help. MRS. L. FULLER Science for the Blind

Popular Electronics is available to the blind and physically handicapped on magnetic tape from SCIENCE FOR THE BLAND, 221 Rock Hill Road, Bala Cynwyd, Pa. 19004. The magazine is read onto tape by volunteer readers with the publisher's permission and is intended solely for the use of the blind and the physically handicapped. If you know someone who could use this service, please pass the word.

"Simplicity + Dwell Meter" (August 1969). In the Parts List, the currentregulator diode should be listed as 1N5299 not 1N5899. The schematic diagram is correct.

smooth

RCA WP-700A, 702A, 703A and 704A constant voltage dc power supplies are all solid-state. A negative feedback circuit maintains constant output voltage with low ripple — regardless of varying line. In fact, at rated load, these supplies are so smooth that "they hardly cause a ripple."

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*Optional Distributor Resale Price.

WP-702A: Siamese Twins of WP-700A, but electrically isolated \$73.00* (five or more) \$87.00* (less than five)

For further information write: RCA Electronic Components, Commercial Engineering, Department K-133W, Harrison, N. J. 07029.

Look to RCA for instruments to test/measure/view/monitor/generate





To obtain a copy of any of the catalogs or leaflets described below, simply fill in and mail the coupon on page 15.

The 1970 edition of the Heathkit Catalog, featuring the world's largest selection of electronic kits, is now available from the Heath Company. The catalog lists more than 300 kits for every budget and area of interest within its 116 pages, 66 pages of which are in full color. Among the items listed in the catalog are hi-fi/stereo components; ham radio equipment; marine gear; test, service, and laboratory equipment; CB radio; and other special interest items. Items of general interest include color and monochrome TV receivers, electronic organs, clock radios, intercoms, and automotive kits.

Circle No. 75 on Reader Service Page 15

Microphones and their accessories are the theme of Catalog No. 2520 now available from the Turner Company, Inc. Listed and described in the new catalog are some 25 or more microphone models-stand-mount, hand-held, lavaliers-for every application from mobile and fixed communications, to hifl, general-purpose and paging. The accessories listed include microphone stands, uprights and flexible arms, replacement cables. line transformers, remote amplifiers, windscreens, and replacement cartridges for microphones. In addition, the catalog tells how to choose a microphone, describes the different types of microphones (cardioid and dynamic), and explains the meaning of frequency performance and output levels.

Circle No. 76 on Reader Service Page 15

Allied Radio Corp. is making available, free on request, their 552-page catalog, No. 290, for 1970. The catalog lists and describes with detailed specifications general-interest and specialty items. Among the general-interest items listed are hi-fi gear, including a large selection of tape recorders; color and monochrome TV receivers; electronics and associated hobby kits; and auto receivers and tape players. Other items include ham and CB equipment; short-wave receivers for the SWL; public address systems; and test instruments. There is also an extensive listing of electronic parts and books.

Circle No. 77 on Reader Service Page 15



CIRCLE NO. 43 ON READER SERVICE PAGE

POPULAR ELECTRONICS READER SERVICE PAGE

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CIRCLE NO. 44 ON READER SERVICE PAGE

Ibrary

QUIET

HANDBOOK OF ELECTRONIC METERS: THEORY AND APPLICATION

by John D. Lenk

Manufacturers of meters provide instructions on the operation and circuit theory of their particular instruments, but rarely do these instruction manuals give any applications data describing the many uses of meters. This book attempts to rectify the oversight. The opening chapters give a simplified presentation of the principles and characteristics of meters, as well as a brief description of meter accessories. Beyond the first three chapters, the book becomes a guide for experienced technicians and engineers. In detailing the greatest number of meter applications available, it covers a full range of practical solid-state and IC data.

Published by Prentice-Hall, Inc., Englewood Cliffs, N.J. 07632. Hard cover. 130 pages. \$10.95.

FELL'S GUIDE TO OPERATING SHORTWAVE RADIO

by Charles J. Vlahos

This is a book written to fill that gap between the position of knowing absolutely nothing about shortwave and the decision to become a ham, or SWL, or even a CB'er. In other words, this is an "introduction" to shortwave radio and to a small extent is similar to the COMMUNICATIONS HANDBOOK published annually by the Editors of this magazine. The author's enthusiasm is contagious and we can recommend his book as a safe guide in making those first steps into shortwave.

Published by Frederick Fell, Inc., 386 Park Avenue South, New York, N.Y. 10016. Hard cover. 180 pages. \$4.95.

UNDERSTANDING AND USING UNIJUNCTION TRANSISTORS

by Stu Hoberman

This book explains in detail the construction, operation, and characteristics of the unijunction transistor (UJT). It describes the various applications of UJT's, and typical circuits in which they are used. Separate chapters are devoted to the use of the UJT in oscillator circuits, voltage control devices, time delay and flasher applications, and in sensing circuits.

Published by Howard W. Sams & Co., Inc., 4300 West 62 St., Indianapolis, Ind. 46268. Soft cover. 96 pages. \$2.50.

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Grantham School of Engineering offers an educational program leading to the Degree of Associate in Science in Electronics Engineering – the ASEE. Beginners in electronics must take resident classes in Hollywood, Calif., or in Washington, D.C. However, experienced electronics technicians may complete the entire degree program by home study, except for the final two weeks which must be taken in Hollywood. There are three different ways the program is offered, referred to as Daytime-Residence, Supplemented-Correspondence, and Home Study. Each method is discussed below.

Daytime-Residence

The Daytime-Resident program is designed for *beginners*. You may enroll in Hollywood or Washington. Classes meet five days per week, and each semester is 16 weeks long. Three semesters are offered each year. Upon satisfactory completion of the five semester program (about 20 months), you are awarded a Diploma in Electronics Engineering Technology. Then, to complete the requirements for the ASEE Degree, you must attend the associate-degree seminar—a two-week period of review, consultation, and evaluation.

This seminar is held, for Hollywood and Washington students, at the main School in Hollywood. For Washington students the School pays the round-trip (to and from Hollywood) airline transportation charges, so that all the graduating students in both schools may participate in each seminar together.

For those who wish to continue their engineering studies beyond the ASEE Degree level, Grantham offers a BSEE Degree program in Hollywood. The Grantham ASEE Degree or other equivalent background is prerequisite to enrollment in the BSEE Degree program.

Supplemented-Correspondence

The Supplemented-Correspondence program is designed for beginners. You take the correspondence lessons from the main school in Hollywood, but the supplementary resident classroom and laboratory sessions, one evening per week, may be taken in either Hollywood or Washington. The main part of the program is divided into five semesters, each semester being slightly less than six months long. so that you normally complete the five semesters in 21/2 years. Upon completion of this five-semester program, you are awarded a Diploma in Electronics Engineering Technology. Then, to complete the requirements for the ASEE Degree, you must attend the associate-degree seminar - a two-week period of review, consultation, and evaluation - in Hollywood, the same as is explained under "Daytime-Residence" above. Seminar round-trip airline transportation for Washington students is paid by the School.

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November, 1969

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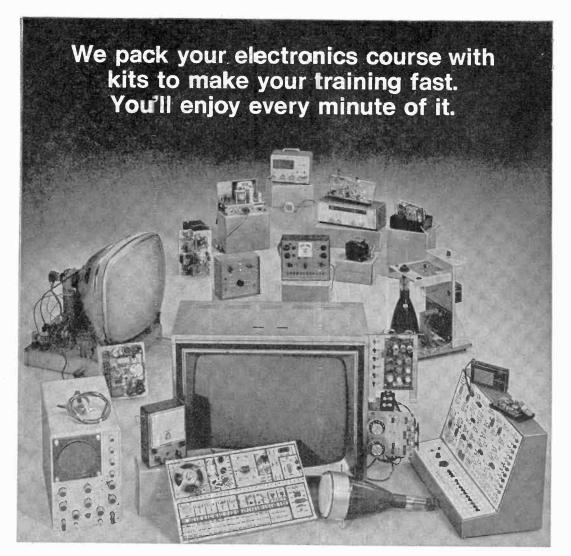
This accredited ASEE Degree program offered to experienced technicians, begins with a review of electrical and electronic principles and systems, and then continues with applied engineering mathematics (including algebra, trigonometry, analytic geometry, and calculus), classical and modern physics, technician writing, computer systems, electrical network design, and semiconductor circuit design.

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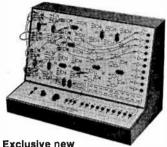
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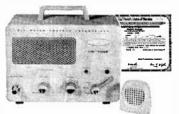




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Additional information on products covered in this section is available from the manufacturers. Each new product is identified by a code number. To obtain further details on any of them, simply fill in and mail the coupon on page 15.

CURVE-TRACER INSTRUMENT FOR SCOPES

The new EICO Model 443 transistor-diode curve tracer makes possible direct readout of semiconductor characteristics on a general-purpose oscilloscope. Diodes, rectifiers, and power signal transistors are tested with this

versatile instrument. Diode and rectifier characteristics that can be displayed include forward voltage, forward current, reverse current, and peak inverse voltage. Transistor tests include those most commonly available with



lab-type instruments of this nature. A special matching switch allows comparison and matching of sets of transistors. The Model 443 has solid-state, printed-circuit board construction; dual transformers for isolation and safety; flashing light that indicates when high voltage is on the diode test terminals; built-in scilloscope voltage calibrators; and terminals for connecting external test sockets.

Circle No. 78 an Reader Service Page 15

DUAL-PURPOSE TAPE ERASER

The Audiotex Model 30-140 is a dual-purpose magnetic tape eraser designed for hand-held



and table-top use. It accommodates both open reel and cassette tapes. As a table model, the 30-140 is the only tape eraser available with a reel post. Hand-held, it is passed over the tape reel in a circular motion. A mo-

mentary-action switch and cord are included.

Circle No. 79 on Reader Service Page 15

ULTRASONIC CLEANER

An ultrasonic cleaner with a 5"-square by 3" deep tank is currently being offered by Esterline Angus as the Model EA 12. The tank size is ideal for cleaning inking components for pen recorders, precision parts, electronic components, optical equipment, lab glassware, surgical and dental instruments, jewelry, and other items. The cleaner can be

operated with common household detergents or with solvent cleaners such as Dow Chemical chlorothene nu or Dupont Freon T-WD. Cleaning power is 50 watts.

Circle No. 80 an Reader Service Page 15

SOLID-STATE AM/FM STEREO RECEIVER

Professional features not often found in economy-priced receivers are standard with the Kenwood KR-77 solid-state AM/stereo FM receiver. The KR-77's FM tuner, for example, incorporates two FET's, a four-gang tuning

capacitor, and two IC FM i.f. stages which combine to offer a 1.9-µV sensitivity, 45-dB selectivity (alternate channel), and a capture ratio of 2.5



dB. The KR-77 permits full amplifier flexibility, with left and right channel preamplifier outputs and corresponding power amplifier inputs to permit separate operation of any amplifier and to make possible simple connection of a multi-channel system. The most often used receiver functions—interstation muting, low and high filter circuits, and loudness control—are available via handy centerpanel keyboard switches. Front panel jacks provide easy access for dubbing/tape record and stereo headphones.

Circle No. 81 on Reader Service Page 15

UHF/FM MONITOR RECEIVER

The "Reporter" six-channel monitor receiver made by *Unimetrics, Inc.*, is now available in a UHF/FM ultra-high-band model. Called the



"UHF/FM Reporter," its frequency ranges include any six crystal-controlled FM channels in the ranges 450-455 MHz, 455-460 MHz, 460-465 MHz, and 465-470

MHz. The 20-transistor/8-diode circuit is designed for extremely low battery drain and miniature size (only 8" × 6" × 2" for the overall receiver). The dual-conversion receiver is ruggedly built to permit heavy-duty use in truck, car, home, boat, or office. Optional accessories include a 117-volt a.c. power supply, external speaker jack provision, and "UNITONE" encoders and decoders.

Circle No. 82 on Reader Service Page 15

AUTOMATIC TURNTABLE

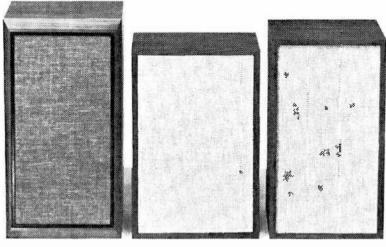
The first new Dual record changer to be introduced in four years by *United Audio Products*, *Inc.*, is the Model 1219. The Dual 1219 features a 12" platter and 8%" tone arm, and a number

of design and operating features unique to automatic turntables. A mode selector shifts the tone arm base down for playing one record and shifts it up for playing a



stack (the tone arm is parallel to the record in single play, shifted to parallel the center of the stack in multi-play). A four-point gyro-

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The Sound of Excellence

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CIRCLE NO. 17 ON READER SERVICE PAGE

PRODUCTS (Continued from page 22)

scopic ring-in-ring gimbal suspension, a new motor designed for high starting torque, a synchronous element that locks running speed into line frequency, and a pitch control which allows each speed to be varied up to 6% are standard items. Also, an extra degree of precision has been incorporated into the antiskate system, and separate scales for each stylus type are provided.

Circle No. 83 on Reader Service Page 15

3" FIELD SERVICE OSCILLOSCOPE

The Model LBO-31M oscilloscope has been designed by Leader Instruments Corp. for field use in addition to multi-channel monitoring applications in the laboratory. The chassis



is narrow, measuring only 4" in width, to allow many scopes to be placed side by side for maximum density and to occupy minimum space on the home work bench. The bandwidth of 1 MHz makes the LBO-31M suitable for use in the videofrequency region. Technical specifications-vertical axis: 80 mV peak-to-peak/cm at 1 kHz sensitivity and -3 dB, 3

Hz to 1 MHz frequency response; horizontal axis-2.5 volts peak-to-peak/cm sensitivity, 10 Hz to 400 kHz sweep frequency, and -3 dB, 3 Hz to 400 kHz frequency response. The scope requires a 105-125-volt a.c., 50-60-Hz power input and consumes only 40 VA of power. Overall dimensions and weight are 12" × 7" \times 4" and 11 lbs.

Circle Na. 84 on Reader Service Page 15

CASSETTE TAPE DECK

"Techno-Built" precision tape heads, a hysteresis synchronous drive motor, and an illuminated dual-level indicator are features of the TEAC Corp. of America Model A-20U stereo

cassette deck. The A-20 also has easy pushbutton loading and unloading, separate input and output controls for each channel, and a unidirectional microphone. Standard deck items in-



clude an editing control, headphone monitoring jack, and three-digit reset counter. Technical specifications-Wow and flutter: 0.2% or less; signal-to-noise ratio: 45 dB or better; frequency range: 60-10,000 Hz; 0.3-mV for 500ohm microphone and 0.5-volt for 120,000-ohm line inputs; 0.5-volt for 50,000-ohm line and 0.1-mW for 8-ohm headphone outputs.

Circle No. 85 on Reader Service Page 15

ANTENNA TRIPODS

Three new gold tripods to facilitate home antenna installation are available from GC Electronics in 2', 3', and 5' sizes. Each of the tripods has an attractive gold baked-enamel finish and is supplied with six lag screws and three tar patches for sealing and mounting holes. The 3' and 5' towers have heavy-duty 14"-diameter legs and a tilt-up feature to facilitate antenna installation

Circle No. 86 on Reader Service Page 15

HIGH-POWER AM/STEREO FM RECEIVER

Instant-acting electronic protection circuits and an electronically regulated power supply are two of the many features available in the new H.H. Scott, Inc., Model 386 AM/stereo FM receiver. The i.f. section employs a crystal



lattice filter that eliminates the need for periodic alignment. An IC in the AM front end features pre-tuned multi-

pole filter for optimum AM fidelity, while another IC in the multiplex section provides greater reliability and a 40-dB stereo separa-tion. Technical specifications—continuous output power: 42 watts/channel at 4 ohms, 35 watts/channel at 8 ohms: selectivity: 40 dB: frequency response: 20-20,000 Hz, ±1 dB, hum and noise: -65 dB; crossmodulation rejection: 80 dB; usable sensitivity: 1.9 μ V; tuner stereo separation: 40 dB; capture ratio: 2.5 dB; signal-to-noise ratio: 65 dB; phono sensitivity: 3.6 mV.

Circle No. 87 on Reader Service Page 15

SUPER-POWER INSTRUMENT LOUDSPEAKER

There is no truth Stretching when Jensen Sound Products claims that their new Model SMI-285 Lifetime Guaranteed Electronic Mu-

sical Instrument Loudspeaker is "Super" high power. Rated at 250 watts, the new 18" speaker features a massive 12½-lb. magnetic structure, using a highly efficient 3%-lb. DP-Alnico-5 magnet to provide all that power. A precisionwound 4" voice coil is fabricated on a special high-power bobbin. A large aluminum



dome provides perfect balance and crisp, clean highs. Solid brass notes are achieved with a laminated, reinforced, flexible edge suspension, and the rugged cast construction of the housing maintains precision line-up and protects against damage from rough handling.

Circle No. 88 on Reader Service Page 15

NEW CONCEPT COMMUNICATIONS RECEIVER

The Mark '6' Scan-O-Matic made by Sonic Industries, Inc., represents the newest concept in communications receivers today. The Mark 6 is much like any other well-designed basic receiver capable of monitoring police, fire, and other public service transmissions. It can monitor up to six crystal-controlled channels. But "Scan-O-Matic" is what sets the Mark 6 off from all other receivers. This special fea-

(Continued on page 118)



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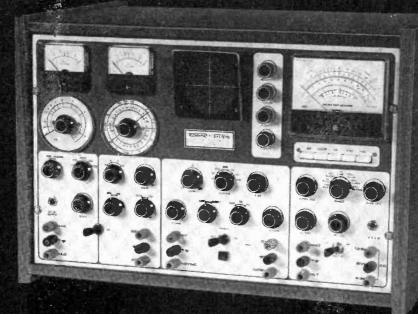
and positive or negative ground operation without internal wiring changes, featuring reverse polarity protection. There's a PA/CB switch with adjustable volume. And the illuminated channel selector and "S" meter makes even night transmission easy. Beautiful, with all silicon transistor, F.E.T. and integrated circuit. It uses 12 volt DC; AC adapter available. Meets FCC requirements. It even comes with its own mounting bracket.

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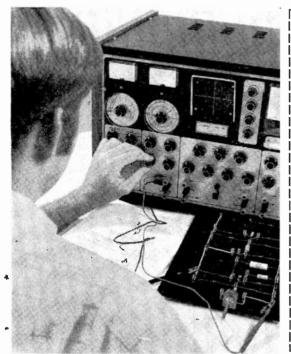
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CIRCLE NO. 40 ON READER SERVICE PAGE

WEIGHS LESS THAN 5 OZ.



BY HOWARD F. BURGESS

N THE 1700's the naive and superstitious believed that "air electricity" could influence both mind and body. A crime committed when a strong dry wind was blowing was believed to be caused by the wind, and some judges in Europe were more lenient during these windy seasons.

Yesterday's fantasy is today's scientific fact. The most exciting research programs develop when an old superstition is found to have some scientific truth. Many, such as the riddle of "air electricity," are being put together one bit at a time. Like a jig-saw picture puzzle some pieces fall into place. Others which can be recognized as part of the picture must wait on a piece that is still missing.

Although there are still some important bits to find, the picture of "air electricity" is becoming clear. We now know beyond a doubt that it can do much to control the mind and body of man. It may be the secret of controlling disease.

In 1921 Frederick Dessauer, of Germany, recognized enough pieces of the puzzle to make a start and by 1931 a picture was beginning to take shape from the data he had been able to collect. He was convinced that "air electricity" was charged air particles which surround us at all times. He had also found that if the negative particles exceed the positive particles, a condition was created which was beneficial to both mind and body. Harmful effects were

found when the positive particles exceeded the negative.

Since the early work of Dessauer, many men and many laboratories have added bits to the picture. Researchers have found that air electricity is really air ionization and the results depend upon whether the majority of the ions are positive or negative in polarity. The negative ions are the good guys and the positive ions seem to be the bad guys.

What is an lon? "Ion" is a short name for a very small piece of matter. Ions are usually measured in millimicrons which are one thousandth of one millionth of a meter. Although small, the physics involved is quite complicated. However, for present purposes the explanation can be quite simple.

An ion is a molecule or group of molecules that has become electrically charged as a result of gaining or losing an electron. A "negative ion" is one which has gained an electron. A "positive ion" is one which has lost an electron. Ions are created in many ways. Any force which can dislodge an electron from an uncharged molecule will create two ions. The molecule which loses the electron becomes a positive ion and the molecule which picks up the wandering electron becomes a negative ion. When ions of opposite polarity collide, another exchange takes place and they are neutralized.

Nature is an endless source of ions. Energy from outer space such as X-rays, ultraviolet, and cosmic rays create ions. Radioactive material in the soil also contributes to the supply. Other natural events such as thunder storms, rain and snow add their effect. Even the wind and the moon have their part in the story.

Air is composed of several gases including oxygen, nitrogen, carbon dioxide and others in lesser amounts. Air also carries varying amounts of pollution in the form of microscopic particles of anything that man can dump into the atmosphere. Water vapor also has a major part in air ionization.

Air ions seldom consist of just one gas molecule. An ion generally consists of a cluster of gas molecules which are sometimes grouped around a water particle or air pollution material. These clusters are classified according to size. Small ions may consist of 3 to 8 molecules and are capable of rapid movement. They are somewhat important for their effect on man. The intermediate sized ions may have several hundred molecules. They move slower than the small ions, and have the greatest effect on living things. Ions classed as large may contain several thousand molecules. They move very slowly and are generally related to air pollution.

For research work ions of the small and intermediate size are desired. These can be generated artificially in several ways. Radioactive sources are very good ion generators but are difficult for the ordinary experimenter to obtain. Ions can also be generated by use of a high voltage applied between special electrodes. Another source is a simple electric heating element at higher than normal temperature. Generators usually require some type of electric filter to remove ions of the unwanted polarity.

Ions are disappointingly short-lived. After generation they will travel only a short distance before being neutralized by another particle.

lon Effects. Many events generally accepted without an explanation are the results of natural ions. The oppressive feeling before a thunder storm that is felt by both men and animals is due to the predominance of positive ions ahead of

a storm. The oppression may take the form of headaches, rheumatism or respiratory attacks. The fresh, wonderful feeling that follows a storm comes from the high level of negative ions that follows a storm front. A misty rain of small droplets usually raises the positive level while a shower of large drops brings up the negative count.

The strong dry winds which occur in some areas will bring up the positive count and may have a marked effect on the temperment of both man and beast. Tests have shown that in areas that have long been noted as health resorts, the ion count many times runs predominantely negative.

In a recent news broadcast from Radio Moscow, Doctor G. Tsitsishvili of the Sanitary and Hygenic Institute stated that it has been determined that the higher you live (above sea level), the longer you live. Although no mention was made of ions, the findings of the doctor are interesting. American researchers have found that the number of negative ions increases with altitude.

As in other things, man alters the level and polarity of ions. Ions produced by nature are generally of the small or intermediate size and are found in clean air. But, air pollution is a factor in generating large ions. Large ions are found in urban areas or where there is air contamination. They are slow in movement. Because they are large and slow, they absorb the smaller but faster ions that collide with them and so reduce any possibility of negative ionization.

Research in air conditioning has shown that ions in an unoccupied room with open windows will be very similar to that outside. If the windows are closed, ion level of both polarities will decrease somewhat. As people begin to occupy the room the number of large ions increases and small and intermediate ions will continue to decrease. The comfort factor will decrease as the number of large ions increase. Forced ventilation through duct work will decrease the ion density but will increase the unbalance in favor of the positive ions. It has been shown that air in close quarters can be kept at the correct temperature and humidity but the occupants will still be quite uncomfortable and distressed if the number of

(Continued on page 114)

MAKE YOUR OWN ION CHAMBER

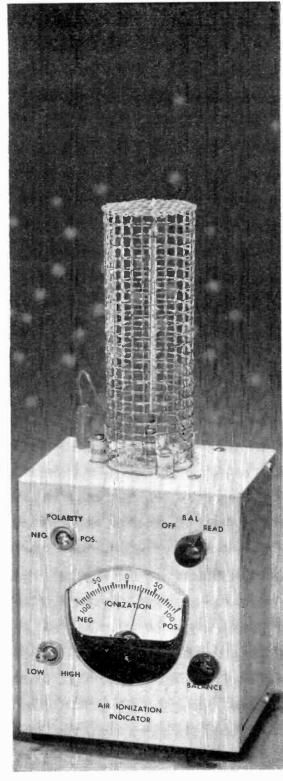
Detect and Measure
Air Jonization Polarity

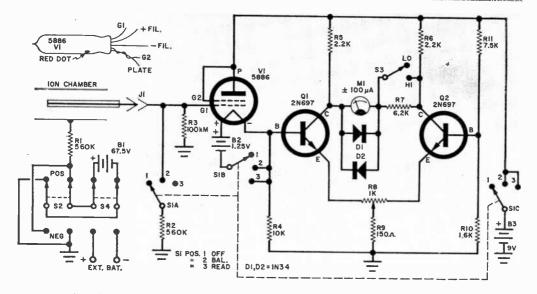
BY HOWARD BURGESS

USEFUL and interesting information can be gained about the effects of air ionization by making measurements under various conditions. A commercial instrument capable of doing this is expensive and delicate but a simple unit can be built which will indicate the polarity and relative amount of ionization present where larger changes are involved.

A home-built ion chamber is used to detect the ions of the polarity which we wish to count. The collected charges are then routed to the grid of an electrometer tube. This tube is a vacuum tube capable of sensing currents far smaller than those which can be detected by an ordinary tube. Currents in this type of instrument can be less than one millionth of a microampere.

The tube was chosen in preference to a field effect transistor because it can take more abuse from the high voltages which may be encountered in this type or work. Luckily the electrometer tube requires only a modest plate voltage. The output of the tube is quite low and requires amplification (transistors Q1 and Q2) to operate the meter, M1.





lons detected by the ion chamber develop a small voltage drop across the very high resistance of R3. This voltage is amplified and operates a transistorized bridge. Sensitivity depends on the value of R3 which must be extremely high to make the circuit effective.

PARTS LIST

```
B1—67½-volt B battery
B2—1.25-volt mercury cell (Mallory RM12R)
B3—9-volt transistor battery
D1,D2-1N34 diode
M1—100-0-100-microampere meter
Q1,Q2—2N697 transistor
R1,R2—560,000-ohm
R4—10,000-ohm
R5,R6—2200-ohm
R7—6200-ohm
R7—6200-ohm
R1—1500-ohm
R10—1600-ohm
R11—7500-ohm
```

R3—100,000-mcgohm resistor* (see text)
R8—1000-ohm, wirewound potentiometer
S1—3-pole, 3-position ceramic switch
S3—S.p.s.l. toggle switch
S2,S4—D.p.d.t. toggle switch
V1—5886 electrometer tube (Allied No. 5886
50F2)
Misc.—Phone tip packs (4), standoff insulator
(3), 1"x3" terminal strip, battery holders,
4"x5"x6" utility box, knobs, hardware, etc.
*Victoreon Instrument Div., 10101 Woodland
Ave., Cleveland, Ohio 44104.

Construction. The unit is housed in a 4" x 5" x 6" aluminum box and construction is quite simple. The small parts are mounted on a $1^{n} \times 3^{n}$ terminal board as in the photo. Switch S1 must be of high quality ceramic. The variable resistor R8 can be a standard wirewound potentiometer; however a vernier or 10-turn potentiometer makes for easier adjustment if you are lucky enough to have one. The only unusual items are the 5886 electrometer tube (V1) and its grid resistor (R3) which has a value of 100.000megohms. Both items are made by the Victoreen Company and are available from sources shown in the Parts List.

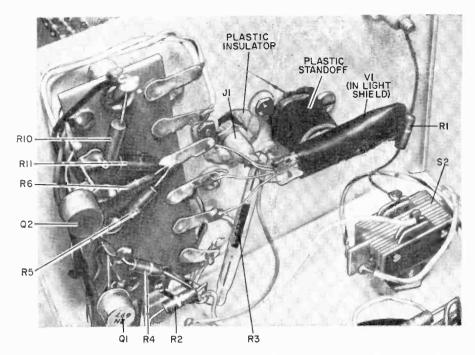
The tube must be covered to protect it from light while it is in operation or the photoelectric effect of light falling on the elements can upset the operation of the sensitive circuit. Only the best of in-

HOW IT WORKS

The ion chamber consists of an outer electrode which is in the form of a screen cage through which air can flow. The inner electrode is a rod in the center of the cage. If a polarizing voltage is connected with the negative to the cage and the positive to the rod, negative ions are attracted to the rod. This produces a voltage across resistor R3. The value of the voltage depends on the number of ions present. This voltage is sensed by the electrometer tube V1 which activates the meter drive circuit. The meter is driven by transistors Q1 and Q2 in a balanced bridge circuit. The zero-center meter will read to right or left depending on whether the charge is positive or negative.

If the polarizing voltage is reversed to the screen cage the center rod will collect positive ions and the meter will read to the right to indicate the level of positive ionization.

The sensitivity of the meter can be reduced by S3 when high field strengths are encountered. The number of ions collected depends somewhat upon the value of polarizing voltage applied to the chamber. For personnel safety reasons the voltage should be kept below 250 volts.



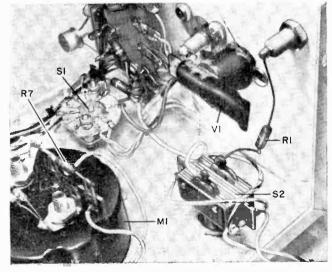
Component installation. Because the presence of light greatly reduces the electrometer tube sensitivity, it must be enclosed in an opaque tape housing. It mounts on a standoff.

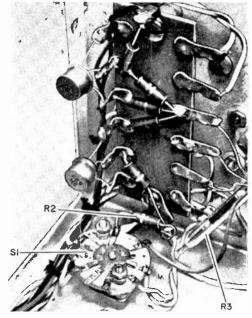
sulation should be used in the grid circuit of the 5886 if full sensitivity is to be obtained. After assembly, all insulation in the grid circuit should be cleaned with alcohol to remove any oils or moisture left by the fingers.

The ion chamber, which is mounted on top of the case, consists of an outer electrode which resembles a small cage 6"

high and 2" in diameter. It is constructed of 4" wire netting as shown in the photos. The bottom is open and is insulated from the metal case by three ceramic insulators. The polarizing voltage of 67½ volts supplied by the battery mounted in the case will be sufficient for most work. For experimental work where higher voltage is required, provision has

Switch S2 reverses the polarity of the ion chamber voltage. Resistor R1 is a safety resistor to reduce current flow in case of accidertal short between cage and case.





Function switch S1 is located near the electronics board. Use a high-quality ceramic switch to prevent leakage around the very high value of resistor R3.

been made to switch to an outside source which can be connected to terminals on the rear of the case.

The cage is electrically above ground by the amount of the polarizing voltage. Resistor R1 has been placed in the circuit to prevent serious shock in case the cage is touched. Voltage to the cage can be removed by the switch in the rear.

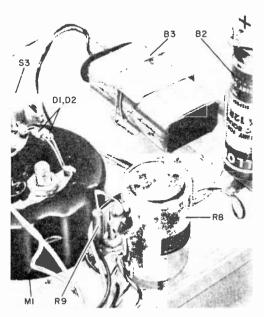
The inner conductor of the chamber is a rod or wire 'n'" in diameter and five inches long. A phone tip is mounted on one end of the rod and plugs into a tip jack. This makes for easy removal for experiments with other types of chambers. The tip jack should be mounted on insulating material of the best quality with a long leakage path. Do not depend on the insulation supplied on the tip jack; cut out a square of quality plastic as shown.

The 10,000-megohm resistor R3 is expensive. To make a substitute, use a ceramic-body r.f. choke, about 1½" to 2" long, with a pigtail at each end. Remove the coil and clean the form thoroughly. Draw a line of Higgins india ink about 1%" wide between the pigtails. Allow the ink to dry completely and never handle the form so as to touch the ink line or otherwise introduce body oils that might reduce the resistance. You may also use a narrow piece of PC board by drilling

a small hole close to each end. Then wrap copper wires through each hole, leaving short lengths to make connections. Solder the wire wraps and draw an india ink line between the two solder joints.

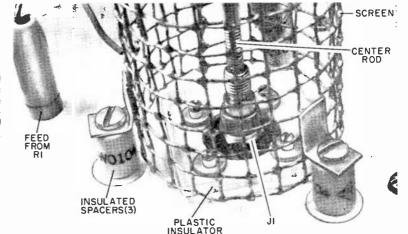
Adjustment and Operation. When the construction is finished and checked, place S1 in the OFF position and S3 in the Low position and install the batteries. When S1 is moved to the BAL position, the unit becomes operational but with greatly reduced sensitivity. The meter should now be adjusted to center zero. If the meter does not adjust to zero with R8 near the center of its range, the value of R10 should be changed to balance the currents of the two transistors. The value of R10 can be raised or lowered a small amount as required.

When the meter zeroes near the midrange of R8 and all appears well, put S1 in the READ position and close S3 by placing it on HIGH. Several seconds may be required for the meter to stabilize. If S2 is in the POS position, the meter will read to the right for any positive charge. Now place S1 in BAL and reverse S2. Return S1 to READ and read the negative charge. Under normal conditions the readings will be small and nearly equal.



Standard potentiometer may be used for R8 instead of 10-turn unit. Diodes mount on meter terminals.

The cage is mounted on three ceramic spacers around the perimeter and is connected via an insulated lead to resistor R1.



Never change S2 except when S1 is in the BAL position.

If a lighted match is brought close to the chamber the results of ionization will be seen. Tobacco smoke blown into the chamber will also indicate ionized particles. If an ultraviolet lamp is available, turn it on the chamber and observe the results.

A small slow-moving fan is useful to force air through the chamber. A fan which stirs the air too violently or which has arcing brushes can generate ions of its own which can make measurements meaningless.

An interesting test can be made on electric heaters. Many heaters generate positive ions and the side effects that go with them. One electric hair dryer tested was capable of pushing the meter off scale from as far away as six feet. The use of such heaters will probably not cause any violent side effects but they have been known to cause drowsiness, fatigue and headaches. Long periods of continued use could be damaging to the general health. Heaters which are rich in ions usually have heating elements which glow brightly.

It has been found that metal duct work in some air conditioning systems creates a positive ion condition by attracting negative ions to the duct walls and leaving an excess of positive ions. The result can be minor respiratory troubles.

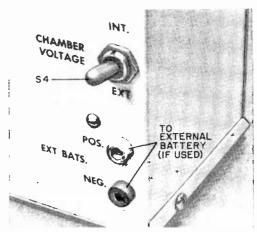
If the meter is placed near an automobile exhaust in a position which will allow the gases to reach the ion chamber large values can be read. In this case both positive and negative ions will

be indicated. These will re-combine in a short time under normal conditions.

Radioactive material will also register on the meter. As an auxiliary use the meter can be used to measure or monitor radioactive conditions or fallout. Other types of chambers may be constructed for use in this field.

If the instrument is operating properly it will be sensitive enough to register movements of your body several feet away. This is a static charge on the body and not ionization. This effect can be reduced somewhat by connecting the meter case to a good ground.

Although this little meter is not as versatile as the sophisticated laboratory models, it will make many interesting measurements and it will introduce you to the fundamentals of what may prove to be one of the secrets of life itself. —30—



Switch S4 is used if it is desired to use a higher voltage for the ion chamber for more sensitivity.

...that's my advice."





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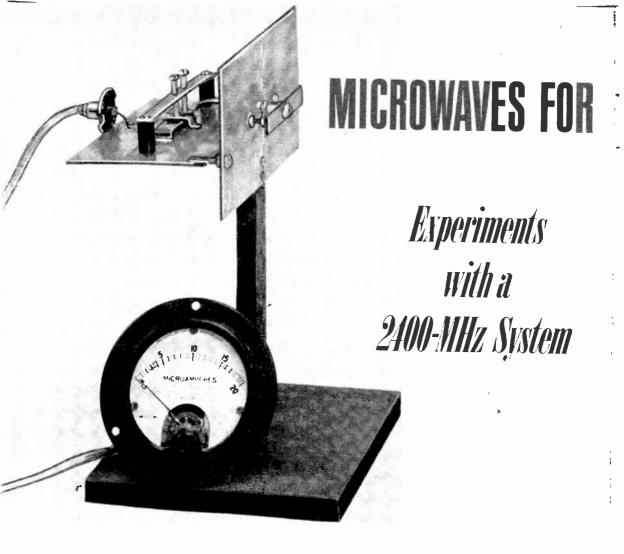
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APPROVED FOR TRAINING UNDER NEW G.I. BILL



AS AN ELECTRONICS hobbyist or ham operator, you have probably built or experimented with receivers and transmitters in all of the usual frequency ranges—up to the top of the 450-MHz ham band. Isn't it about time you tried microwaves? In the past, this has been difficult—special vacuum tubes or solid-state devices that operate in the microwave region were expensive and some of the metal-working (plumbing) required was difficult.

The microwave system described here—including transmitter, antenna, and receiver—can be built at a moderate cost, with a small amount of hardware work. It operates in the S band from 2300 to

2450 MHz; and although the transmitting power and range are very low, building and experimenting with this system will give you a good insight into how microwaves are generated, propagated, and detected. In principle, the system behaves just like its big brothers in telephone and TV distribution.

In most microwave work, it is customary to use a technique known as "stripline" in constructing the tuned circuit. Stripline consists of a lamination of a copper ground plane, a layer of insulation, and then a thin strip of copper to conduct the r.f. energy. Unfortunately, making stripline requires a precision layout and chemical etching, and it is



not easy to tune. The construction technique used in this project results in a tuned circuit that is electrically the same as stripline but it is easier to build and tune. In this case, we use a ground plane, air and nylon for insulation, and ordinary thin copper strips for the r.f. conductors.

Transmitter. The schematic of the transmitter is shown in Fig. 1, while Figs. 2, 3, and 4 show the method of construction. The ground plane is a piece of conventional copper-clad PC board measuring about $4\frac{1}{2}$ " \times 3".

measuring about $4\sqrt{2}^{\circ} \times 3^{\circ}$. The tuned circuit, L1, is made from a piece of thin (about 0.024°) copper %6'' wide \times 1%6'' long. Bend a %'' lip at each end, then make another bend %2'' from each lip. The %'' lips are used to solder the inductor to the ground plane, so that the inductor is 1%6'' long and stands %2'' from the ground plane. Apply heat to the lips and coat the bottoms with solder. Center the inductor on the copper foil of the ground plane and tin the area under each lip. Then solder the inductor to the copper foil. Pick one side of the ground plane to be the front and the other side the rear.

Make up bypass capacitor C1 using a thin sheet of copper or brass $\frac{1}{2}$ " square. Deburr the edges, and drill a small hole in the center to accommodate a small

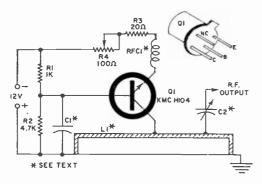


Fig. 1. The transmitter uses a special microwave transistor as a very low-power oscillator.

screw—either metal or nylon. Drill a similar hole through the ground plane at approximately the center of L1 and about $\frac{1}{2}$ of the way from the edge of L1 to the rear edge of the ground plane. If you use a metal mounting screw, scrape away a small area of copper around the ground plane hole so that the screw will not make electrical contact with the ground plane.

Cut a piece of mica (a power transistor mounting insulator will do) or a small piece of 3-mil fiberglass slightly larger than the plate of C1. Mount the capacitor on the insulator and secure it in place with one edge of C1 parallel to the length of C1. If you use metal mounting hardware, use an ohmmeter to make sure the capacitor plate is not making contact with the ground plane.

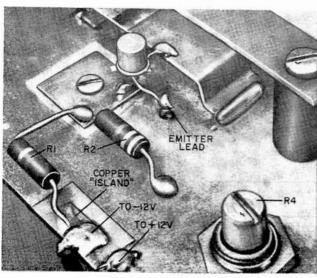
PARTS LIST TRANSMITTER

C1, C2--Sec text
L1--Sec text
(II--Sec text
(II--Sec text)
(II--KMC II 104 high-frequency transistor*
R1--1000-ohm, ½-watt resistor
R3--4700-ohm, ½-watt resistor
R4--100-ohm potentiometer
RFC1--Five turns of #30 DSC wire, on ½16"
diameter
Misc.-1½" x 3" conventional copper-clad PC
board (2), thin brass sheet for C1 and C2, nylon screws and nuts (available at hobby shops),
thin copper strip for L1 and antenna, metal
hardware, solder, etc.

*An H104 transistor is available from Rayville Associates, Mr. S. Nelson, Parker Rd., RD 2, Long Valley, N.J. 07853, for \$5.

Using Fig. 2 as a guide, drill a hole at the rear corner of the ground plane large enough to accommodate the shaft of potentiometer R4. Scrape out an island in the foil near the R4 shaft to form a terminal for the 12-volt connection, one end of R1, and a lead to R4.

To make antenna coupling capacitor C2, cut a piece of thin brass so that it is about 1 inch long and very narrow except for a $\frac{1}{2}$ " square tab at one end. This is the capacitor plate. Glue a mica or fiberglass insulator sheet, slightly larger than the capacitor plate, to the top of L1, centered about $\frac{1}{2}$ " from the end of L1 (see Fig. 3). Drill a hole in the ground plane about half way between L1 and the front of the ground plane large enough to accommodate a nylon screw. Then following Fig. 3, affix the thin end



Fib. 2. The emitter of Q1 couples to RFC1 via an insulated wire fed through a hole in the PC board. Use a sharp blade to create an insulated "island" for the -12-volt feed. Don't forget the mica insulator underneath bypass capacitor C1.

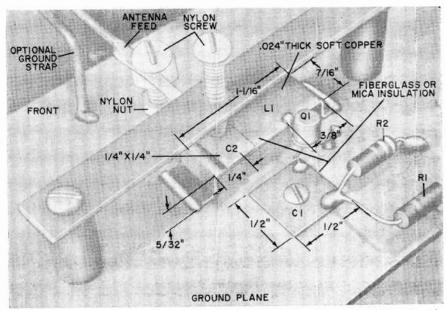


Fig.3. Mechanical construction of the transmitter. Antenna coupling capacitor C2 has its thin lead wedged between the nylon support screw head and a nylon nut. Use a piece of scrap PC board to make the tuning screw support bridge, using a metal screw to tap the C2 hole.

of the capacitor lead to this screw using a nylon nut and making sure that the plate end of the capacitor fits directly over the insulator on L1.

Using a piece of scrap plastic and standoff insulators, make a "bridge" over L1 as shown in Fig. 3. Mark off a point directly over the plate of C2 and drill a small hole in the bridge. Use a metal screw to tap the hole and insert a nylon screw (having the same pitch) in the bridge so that it touches the capacitor plate and can be used to adjust the spacing between the capacitor and L1.

Note where the collector of transistor Q1 is to be soldered to L1 as described in the next paragraph. Directly under where the emitter lead will be, drill a small hole to accommodate a length of insulated wire.

Using a heat sink on the transistor lead to protect the device from heat, solder the collector lead to a point %'' from the end of L1 and the base lead to the metal bypass capacitor C1. Bend the emitter lead toward the hole drilled in the preceding paragraph. The fourth lead on the transistor is not used and can be cut off short. Connect R1 and R2 into the circuit.

On the underside of the ground plane

(see Fig. 4), connect one end of R4 to the -12-volt terminal. Connect R3 to the other end and the rotor of R4. Choke RFC1 is made by winding five turns of number 30 DSC wire around any $\frac{1}{16}$ form. You can impregnate it with coil dope or wax to make it retain its shape. One end of RFC1 is fed through the hole in the ground plane and soldered to the transistor emitter lead. The other end is soldered to R3.

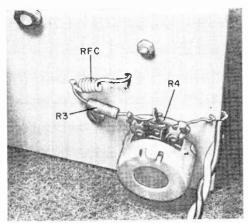


Fig. 4. Underside view of the transmitter showing the location of the RFC and R3. One end of the RFC feeds through a hole and contacts Q1 emitter.

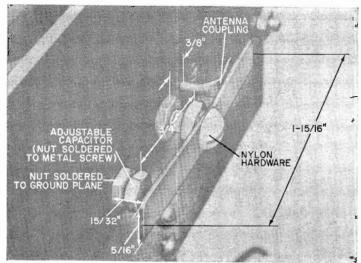


Fig. 5. The antenna coupling (from C2 on the transmitter) is made by creating a flat on one end of a piece of soft-copper bus bar. To install it, rotate the antenna by 90-degrees. After installation, the antenna coupling should be close, but not touching the antenna proper.

Antenna. The ground plane for the antenna is the same physical size as that used for the transmitter. To make the actual antenna (see Fig. 5) cut a piece of thin copper sheet $1^{15}\%6'' \times \%6''$. Drill a hole in the center of the copper and in the center of the ground plane so that the antenna can be mounted using nylon hardware.

The antenna is tuned by a small nut soldered to the end of a conventional 6-32 metal screw. Drill a hole in the ground plane 34" from the antenna mounting screw hole and on the same center line. Solder a 6-32 nut to the copper foil of the ground plane so that it is centered on the hole just drilled. Thread a 6-32 metal screw an inch or so long into the nut from the back so that it protrudes to a maximum on the antenna side. Put a metal nut on the screw so that the nut is flush with the end of the screw. Solder the nut to the screw and file the end surface flat. This screw is now a variable capacitor, with the flat end of the nut-screw combination capable of being adjusted with respect to the antenna (when it is installed). On the other side of the antenna mounting hole. about %" away and still on the antenna center line, drill a hole large enough for a piece of round copper bus bar. Trim the copper foil away from the perimeter of the hole to avoid accidental grounding when the antenna feed is installed.

Take a piece of soft-copper, round bus bar and, using a hammer, make a ¼" ×

4" flat on one end. This will be the antenna feed.

Solder a small L-bracket at each corner of the front of the transmitter ground plane. Hold the antenna ground plane at right angles to the transmitter and adjust the height so that the antenna feed hole is in line with the slim metal lead on the antenna coupling capacitor (C2) secured in the nylon screw.

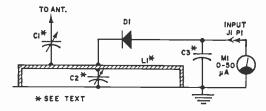


Fig. 6. Because of the very high frequencies used, a special microwave diode is used in the receiver. The diode can be inserted with either polarity to L1, as long as it matches the polarity of meter M1.

PARTS LIST RECEIVER

C1, C2, C3-Sec lext D1-PD0820 diole* J1-Phono jack

L1-See text

P1—Coaxial phono plug

Misc.—4½" x 3" conventional copper-clad PC

board (2), thin brass sheet for C1, C2, and C3, nylon screws end nuts (available at hobby shops), thin copper strip for L1 and antenna, metal hardware. solder, etc.

*A PD0820 Schottky barrier diode is available from Parametri; Industries, Inc., 742 Main St., Winchester, Mass., 01890 for \$2.75.

Attach the antenna ground plane to the L-bracket so that a continuous ground is made from the transmitter ground plane to the antenna ground plane.

Once the two ground planes are fastened, pass the antenna-feed bus bar through its hole in the antenna ground plane so that the tab is approximately ${}^{1}\%_{2}{}^{"}$ from the ground plane and bent out toward the end of the antenna. Be sure the feed line does not touch the ground plane. Cut the other end of the bus bar so that it can be soldered to the end of C2 in the transmitter. Rotate the antenna tuning capacitor (on the antenna ground plane) so that the soldered-on nut is touching the nut on the ground plane.

Slide, or thread, the actual antenna onto an inch-long nylon screw until it touches the head of the screw and secure the antenna in place with a nylon nut. Place another nylon nut on the screw and place the screw into its hole so that the antenna is 15/32" off the ground plane (about 1/5 of a wave-length). Secure the antenna in place with a metal (or nylon) nut at the rear of the board. Position the antenna feed so that it is close to but not touching the antenna.

The antenna-transmitter combination can be mounted on a wood base. Connect a wire to the -12-volt terminal "island"

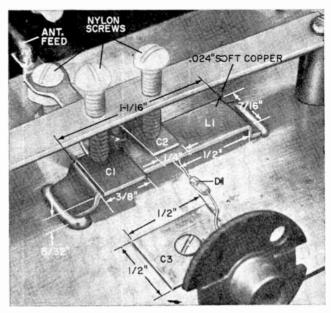
and another wire for the +12-volt connection to the main foil of the transmitter ground plane.

Receiver. The schematic of the receiver is shown in Fig. 6. Inductor L1 and bypass capacitor C3 are fabricated just as they are for the transmitter and are shown in Fig. 7. The receiver has a tuning capacitor (C2) made up of a piece of thin brass sheet, $\frac{1}{4}$ " wide, bent so that a small lip can be soldered to the ground plane at the center of L1. Another lip on the top straddles L1 but is separated from it by an insulator. The antenna coupling capacitor C1 is similar to that in the transmitter. Nylon screws for adjusting C1 and C2 are supported in the plastic bridge over L1.

Diode D1 is soldered to the center of L1 (either polarity) and to the bypass capacitor. A phono jack with the shield pin soldered to the ground plane and the center pin connected to the bypass capacitor serves as a meter connector. The antenna for the receiver is constructed in the same manner as that for the transmitter.

A 50-microampere meter is used to measure the signal level of the rectified signal output from D1. The meter can be connected through a length of twinconductor cable and a suitable plug and placed for easy viewing.

Fig. 7. Construction details of receiver. Strapline L1 is the same as used in the transmitter. Be sure to install mica insulators under capacitors C1, C2, and C3. Receiver uses the same antenna construction as that used with the transmitter.



HOW IT WORKS TRANSMITTER

In the tuned circuit, the r.f. is bypassed by capacitor C1 and bias is determined by the network of R1 and R2. The base of Q1 is 180° out of phase with the oscillator collector r.f. voltage. This is the necessary condition for oscillation. The emitter of Q1 is r.f. decoupled from the power supply by RFC1 and the transistor current is determined by the setting of potentiometer R4. There is an internal in-phase feedback from collector to emitter.

The tuned circuit. L1, is a half-wave line grounded at both ends. At one instant, the center of L1 (collector connection) goes r.f. positive, while the base and the ground plane are negative. One half-cycle later, the polarities of both points are reversed. If the Q (figure of merit of L1 and the circuit as a whole is sufficiently high, the circuit will oscillate at a frequency determined by the physical dimensions and distance from the ground plane of L1.

The amateur band of 2300 to 2450 MHz has been picked for this experimental microwave oscillator because that region is approved for experimental amateur purposes. Changing the dimensions of L1 and its proximity to the ground plane permits tuning the transmitter to any fre-

quency within the band.

Increasing the emitter current by varying R4 increases the transmitter output power. Resistor R3 is a safety resistor which prevents OI from burning up from too much current. Trimmer capacitor C2 couples the r.f. output to the antenna.

Operation. To tune the receiver to the transmitter, place the two units a short distance apart on the workbench, with the antennas facing each other. Supply power to the transmitter from a 12-volt d.c. source being careful to observe the correct polarity. Set R4 between half and % resistance. The receiver meter may or may not show an indication of output. Gently adjust C2 on the receiver, looking for a meter indication. Tuning may be sharp, so be careful. If the two units are close together in frequency, the meter should show a signal somewhere in the course of tuning C2.

Once a signal is detected, adjust the transmitter antenna coupling capacitor C2, the transmitter antenna tuning capacitor and potentiometer R4 for maximum signal. With the transmitter rough tuned, adjust the receiver antenna coupling capacitor C1 and the receiver antenna tuning capacitor for maximum signal. Once these have all been adjusted for a maximum signal, trim the transmitter and receiver coupling tabs for maximum signal and recheck all variable adjustments. As the received signal strength increases, it may be necessary

to separate the two devices by a couple of feet more.

Frequency Measurement. Once the microwave system is operating, a frequency check must be made to make sure that it is operating within the 2300-2450-MHz band. The approach (see Fig. 8) is called "interferometry" and was first used over a century ago to measure the frequency of light waves. (Heinrich Hertz used it in 1887 to measure the first electromagnetic waves ever generated.)

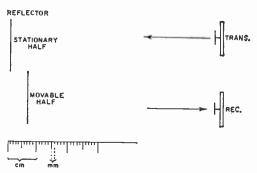


Fig. 8. Basic interferometer is used to measure frequency. As the movable half of the reflector is brought closer to the antennas, a series of "nulls" will be seen on the receiver meter. You measure the distance between nulls to determine frequency.

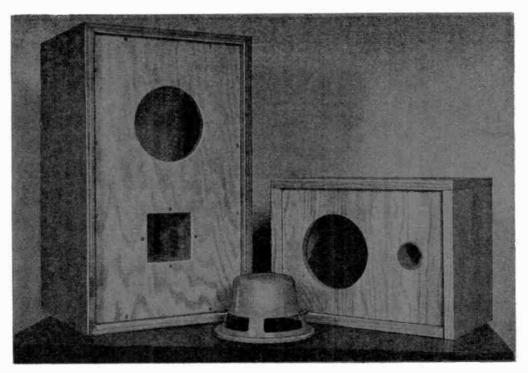
Place the transmitter and receiver side by side so that their antennas are facing in the same direction. Place a ruler calibrated either in inches or centimeters at 90 degrees to the plane of the antennas and about 2 feet in front of them.

Make two flat metal reflectors (several inches on a side) and place them in front of the two antennas at the zero mark of the ruler. The butting edges of the two reflectors should be approximately cen-

(Continued on page 112)

HOW IT WORKS RECEIVER

The tuned circuit of the receiver is similar to that of the transmitter. Trimmer capacitor C1 couples the antenna to the tuned circuit. The rest of the circuit is a conventional crystal detector with a microammeter as the readout. Because of the very high frequencies involved, a Schottky-barrier diode must be used. Unlike conventional diodes, this device is especially made to operate efficiently at microwave frequencies. Capacitor C3 is an r.f. bypass for the meter.



Rally Round the Reflex

BY DAVID B. WEEMS

Proper enclosure tuning gives maximum enjoyment

MAN YOUR sabre saws, reflex partisans, and look to your ports; the much maligned bass reflex is again under attack! Let's start a campaign to take the boom out of the boom box and preserve it from extinction. The hi-fi system you save may be your own!

Few people will argue that the idealized bass reflex enclosure is efficient and that you can readily obtain ear-shattering sound from only a few watts of audio power. The "boominess" of the homemade bass reflex results from either the lack of know-how or the inability to tune the enclosure properly. Any solution to the boominess and size problems will depend upon whether you want to

emphasize transient response or clean low-frequency reproduction and bass efficiency.

All enclosure designers must make a compromise at some point or another in building a bass reflex. Presumably each manufacturer makes the decision which best complements his loudspeakers, but he may have been influenced by what he believes most customers want. If you prefer the satisfaction of doing the job yourself or just want "something differ-

^{1.} The author attempted to answer these questions in his article, "Tune Up Four Bass Reflex," POPULAR ELECTRONICS, July 1968, page 41. Unfortunately, while answering one question he emphasized the puzzling problem of how big to make the enclosure. (Editor)

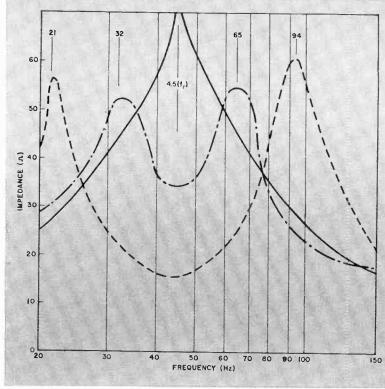


Fig. 1. Graph at left shows speaker's free air impedance (solid line), impedance in 1300 cu in. tuned enclosure (short dashed line), and impedance in 8600 cu in. tuned enclosure (long and short dashed line).

ent," here are some guides to help you. Anyone can use the chart design method, but those with test equipment will want to apply the impedance checks. For easy tuning we shall stick to enclosures with ordinary ports which can readily be adjusted after completion.

Reflex Working Principles. In order to consider enclosure size and other questions, let's take a quick look at how the ported enclosure works. A speaker in a closed box acts like a piston, compressing the air in the box as the cone moves in, lessening the pressure in the box as the cone moves out. Add a port to the box so the pressure can be relieved and you have an adaptation of a Helmholtz resonator. The term "resonator" indicates that the air in the box is resonant at a certain natural frequency. The frequency of the resonance is determined by the enclosed volume of air and the size of the port.

You might think of the air in the box as a large spring, compressing and expanding, with a piston at each end. One piston is the air in the port, and the oth-

er piston is, of course, your speaker. At resonance, both pistons act to compress the spring at once, another way of saying that the two pistons are in phase.

What effect does this have on the speaker? A speaker cone in free air, or in a closed box, vibrates at increasing amplitude as the signal goes down in frequency (below 200 Hz) with the exception of the resonant point where there is a peak in amplitude. Because of this, speakers used in small sealed enclosures are designed with long-throw voice coils so that the voice coil can move over a comparatively large distance and still remain in the region of uniform magnetic field. Such speakers are relatively inefficient.

Speakers designed for large ported enclosures are high in efficiency. When one of these speakers is mounted in a ported box, and the system is working at resonance, the air in the enclosure (being under compression when the speaker moves in and at a partial vacuum while the speaker cone is moving out) acts on the cone in opposition to its movement. This damping reduces cone

travel at resonance, resulting in significantly lower distortion at low frequencies. And the sound output is augmented by the supplementary "piston"—the air in the port. In a properly designed system, port radiation is at a maximum in the frequency range where it is most needed to compensate for natural bass loss

Enclosure Size. Your choice between moderate, large, or colossal enclosure size may depend on whether you want to emphasize low frequency reproduction, transient response, or peace in the familv. The design chart, Table I, shows that the volume determines the tuning frequency for a given size port. Or stated differently, the larger the enclosure, the larger the port for a given frequency. And the larger the port, the greater the radiation from the port—up to a certain point. Beyond that, as the port becomes larger, the enclosure is changed from a box with a hole in it to a box with one side missing and the air behind the speaker ceases to act on the speaker. A useful rule of thumb is to set the upper limit of port size equal to the effective cone area of the speaker. In other words. the box should be no larger than that which allows you to tune the enclosure at the desired frequency with such a port.

Even this may be too large for optimum performance, particularly if you like "tight" bass. The goal then is to choose an enclosure volume that is large enough to yield a port area that is sufficient to be effective without ringing or sounding "mushy."

At the other end of the size range, a ported enclosure that is too small has several undesirable characteristics. Small enclosures demand small ports, and a tiny port may produce more whistling noises than bass notes. Similarly, avoid a slit port unless you are using several of them as an acoustical resistance. If you find that the correct port area for a given situation is only a few square inches, use a duct instead of a simple port.

Another disadvantage of the too small box is the greater tendency to boom. And the low bass notes are cut off. Even ducts can't alleviate these faults satisfactorily.

If you have no audio test equipment, you must resort to design charts. One rule of thumb is to look at the chart and choose an enclosure volume that will require a port area from 30 to 100% of the effective cone area of your speaker for proper tuning (more about this later). The range in square inches is:

Rated Speaker	Port Area
Diameter (in.)	(sq. in.)
8	9 to 28
10	15 to 50
12	24 to 78
15	40 to 133

You can choose your enclosure volume within the limits determined by the port areas given above in relation to the space you have available.

Here is an example. You have an 8-inch speaker that requires an enclosure tuned to 45 Hz. From Table I vou see that the largest recommended enclosure volume is 6 cu. ft., requiring a port area of 28 sq. in. The smallest recommended volume is 3.5 cu. ft. with a port of 11

	TABLE I				RTS
	PC	RT A	REA ((sq in	.)
VOLUME	POR	T FRE	EQUE	VCY (Hz)
(cu ft)	35	40	45	50	60
12	38	52	90	130	0.0
11	34	48	84	120	00
10	30	40	64	90	* *
9	26	36	56	80	0.0
8	24	30	50	74	145
7	18	24	36	50	110
6	12	18	28	36	75
5	9	14	20	30	50
4		10	15	20	36
31/2		*	11	16	28
3	4	٠	8	12	22
21/2	134	¢	*	8	16

The figures given are approximately correct for simple square or circular ports. For rectangular ports (length:width ratio = 4:1) the port should be reduced in area to about 75% of that shown here.

*Use ducts.

**Close port.

sq. in. Note that these values are given for simple ports; ducts must be used for smaller enclosures.

To choose the best enclosure volume for your speaker system, you should use every aid available, including all the information supplied by the manufacturer. With test equipment you can also run an impedance curve for the speaker.

Use of Impedance Curves. The impedance of a speaker is usually rated at 4, 8, or 16 ohms at X Hz. Because the impedance of a voice coil is made up of its d.c. resistance plus both capacitive and inductive reactances, which change with frequency, it varies widely over the audible spectrum.

Figure 1 is a typical graph of impedance vs frequency to show something of the effects of an enclosure on a speaker. A speaker operating in free air has a single low-frequency peak in its impedance curve. The peak occurs at the fundamental resonance point. Put the speaker in a closed box and the peak is still there, but now it's higher in frequency. Port the box and there will be two peaks with the valley between them located at the frequency to which the system is tuned, usually at the free air resonance of the speaker.

Don't try to equate impedance curves with frequency response. As James F. Novak, Senior Design Engineer for Jensen, says, "Any amplifier of reasonably high quality has a damping factor good enough to make it a constant voltage source. With this type of amplifier, the speaker/enclosure response is independent of the impedance variations. If the amplifier were constant current, the response would follow the impedance curve."

Nevertheless, the impedance curve does provide useful information. As mentioned earlier, it tells you to what frequency your system is tuned. And it tells you the frequency of the upper peak, the greatest danger zone for boom in ported enclosures. If this peak occurs above 100 Hz, into the male voice range, announcers and some singers may sound as if their microphones were located in the bottom of a barrel.

The impedance curves shown in Fig. 1 demonstrate the effect of enclosure size on impedance. These curves are for a good quality 8-inch speaker in two enclosures (the smallest and largest recommended sizes), each tuned to 45 Hz, the free-air resonance of the speaker.

It is evident that the small enclosure has the greater effect on the impedance of the speaker. Stated differently, the smaller enclosure is more closely "coupled" to the speaker. The speaker impedance is much lower at resonance in the small enclosure than in the larger enclosure. This indicates that the smaller enclosure is doing more effectively what a tuned enclosure should do—control the cone movement at resonance. But it is impossible to say, just looking at the impedance curves at resonance, what enclosure volume is optimum.

However, you can get some idea of the frequency response range of the two enclosures. The bass output from ported enclosures cuts off at some point below the frequency of the upper impedance peak. Looking again at the curves, you will see that the small enclosure will cut off somewhere below 94 Hz. But the larger enclosure will cut off somewhere below 65 Hz. This difference would be noticeable, particularly in the reproduction of musical tones of large instruments.

Another difference in the impedance curves produced by the large and small enclosures is the frequency span between the impedance peaks. Note that the twin peaks of the large enclosure are much closer together in frequency than the peaks of the smaller enclosure. A useful term of reference in measuring this span is to divide the frequency of the upper peak by the frequency of the lower peak. The quotient is then simply the ratio of the upper frequency peak to the lower frequency peak, a convenient small number. For the enclosures referred to in Fig. 1, the ratio is 2(65/32) for the larger enclosure and 4 (94/21) for the smaller enclosure. The larger the enclosure, the closer together will be the twin peaks and the lower the ratio.

The ratio should fall within certain limits for enclosures of proper volume. For conventional loudspeakers of the 1950's, the limits were between 1.5 and 2.4.2 These low values would result in extremely large enclosures by today's standards. One reason for larger enclosures in those days was that many high-fidelity loudspeakers then had a resonant frequency of about 60 to 70 Hz. It was necessary to use a large enclosure so that the upper peak would be below 100 Hz. For today's low-resonance speakers a more reasonable range might be from about 2 to 3.

^{2.} F. Langford Smith, Radiotron Designer's Handbook, 4th ed.; Radio Corp. of America, 1953, p. 847.

James F. Novak has suggested the precise ratio of 3.13 which he says produces optimum transient response. It yields rather compact enclosures except in the case of very-high-compliance, low-resonance speakers. For the latter, simple rules of thumb do not apply, and some reduction of interior volume will give excellent results with ducted ports even when the peak:peak ratio is somewhat greater than 3.13.

You can use this ratio as a rough guide to estimate whether your enclosure is at the large or small end of the scale for your particular speaker. Listening tests conducted with three enclosures of various sizes with peak: peak ratios ranging from 4 to about 2 indicated that the average listener preferred a ratio of 2 (the largest enclosure) by a wide margin. Several listeners greeted the sound from the box with the peak:peak ratio of 4 with sarcastic comments.

Unfortunately, it is difficult to predict in advance what the ratio will be for a particular speaker/box combination. Novak has described a useful method of estimating the preferred values for his optimum volume enclosures (with a peak:peak ratio of 3.13) by measuring the free-air resonance of a speaker and then measuring its resonance in a "standard" box.3

If you test your present enclosure and find that it is too small or too large by these standards, there are possible cures without major surgery. For the small box, a collar of damping material, such as fiberglass, over the speaker will often help. And for the rare large system that has too much port output, a cloth stretched tightly across the port will usually be sufficient.

If you are designing an enclosure, you should consider its shape as well as its volume because the shape will have some effect on frequency response and impedance. Try to avoid extreme shapes, such as a cube or a pipe. A good general rule is to make each inside dimension slightly different, but limit the longest inside measurement to less than 3 times as great as the shortest. Some manufacturers select precise ratios for these

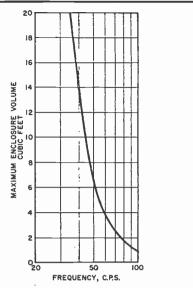


Fig. 2. Graph shows maximum volume vs. frequency for a Helmholz resonator in the shape of a rectangular box with the dimensions in the ratio of 1:1.44:2.08.

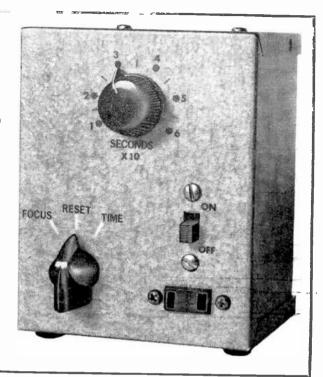
dimensions. For example, John Gilliom at Electro-Voice states that the optimum dimension ratio is 1: 1.44: 2.08 (see Fig. 2). In other words, a typical bookshelf enclosure might have an inside depth of 1 ft., an inside height of 1.44 ft. (about 171/4"), and a length of 2.08 ft. (about 25"). Or the 5 cu ft "universal" enclosure shown in Fig. 3 would have approximate inside dimensions as follows: A 30"; B 21"; C 141/2".

Tuning the Ported Enclosure. Tuning an enclosure to a specific frequency is not very difficult (see Popular Electronics, July 1968, for an easy method), but choosing the frequency may not be so simple. The traditional goal, almost universal in articles and books about ported enclosures, is to tune the box to the freeair resonant frequency of the loudspeaker. Theoretically, the tendency of the speaker cone to undergo excessive movement at resonance would be controlled by the resistive action of the air in the box as described earlier. Although the practice is still followed by some companies, there are certain situations in which poor bass is the result.

(This article will conclude next month with more about tuning the enclosure and instructions for running an impedance curve.)

^{3.} James F. Novak, "Designing a Ducted Port Bass Reflex Enclosure," Electronics World, January 1966.

Build the SCS Darkroom Timer



REPEAT ACCURACY IN A PROJECT FOR THE PHOTO BUFFS

BY FRANK H. TOOKER

AN ESSENTIAL for making clean, crisp photographic enlargements is a dependable means of accurately controlling exposure time. The specially designed instrument described here is continuously variable over the range from 5 to 60 seconds. Its function switch has a FOCUS position for quick and easy setup of the enlarger as desired, or required, for each negative. Its accuracy is better than that of many previous electronic timers because the semiconductor used does not load the RC circuit which determines the timing interval.

The majority of darkroom practitioners are aware that most enlargements are made in the range of 10 to 40 seconds printing time. Intervals of less than 10 seconds make control (manipulation) measures difficult, while those longer than 40 seconds increase the possibility of blurring. Only when considerable manipulation is required is the full 60-second interval actually needed. However, it should be included in every darkroom timer.

The timer's circuit is somewhat susceptible to ambient temperature variations, but far less so than the photographic processes themselves. (Since optimum darkroom temperature is around 68° to 70° F, calibration at that temperature is desirable.) The timer's response to line-voltage transients is negligible. If the test-strip method of determining exposure is used, repeat accuracy is more important than absolute accuracy and the repeat accuracy of this timer has been found to be 3% or better.

Construction. The circuit of the timer is shown in Fig. 1. It is assembled in a $5'' \times 4'' \times 3''$ aluminum cabinet using the front-panel layout shown in Fig. 2. Figures 3 and 4 show the internal arrangement. Power transformer T1 and relay K1 occupy the bottom of the cabinet while the power supply is built on a perforated (or etched) board supported by the transformer mounting hardware. Most of the other electronic components

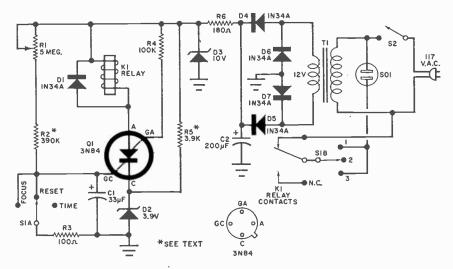


Fig. 1. Designed around a new semiconductor—the silicon controlled switch—this circuit features excellent accuracy and repeatability over the 5- to 60-second range.

PARTS LIST

(1-33µF, 15-volt tantalum capacitor
(2-200-µF, 25-volt tactrolytic capacitor
)1,104-D7-1N3-4A diode
)2-3.9-volt, 1-watt sener diode
D3-10-volt, 400-mW sener diode
K1-2500-ohm relay, 2-ampere contacts, 3-m.1
coil current (Sigma 4F-2500S/SIL or similar)*
(11-3N84 silicon controlled switch
R1-5-megohm, linear taper potentiometer
R2-390,000-ohm (see text)
R3-100-ohm
R4-100,000-ohm

R5-3000-ohm (see text)
R6-180-ohm
S1-2-pole, 3-position, non-shorting rotary switch
S2-S.p.s.t. slide switch
S01-17-volt receptable
T1-12-volt, 10-mA power transformer (Burstein-Applebee No. 18A1387 or similar)
Mise.-5" x 4" x 3" aluminum cabinet, knobs
(2), line cord, terminal strip, mounting hardware and spacers, rubber feet (4), dry transfer lettering, etc.
Adiust armature spring tension, if necessary for

positive pickup.

HOW IT WORKS

The heart of the darkroom timer is a silicon controlled switch (QI). When the cathode-gate of QI is made sufficiently positive, it turns on. When it is made negative—or when the cathode is made positive with respect to the cathode-gate—the switch turns off. Sensitive relay KI, in the anode circuit of QI, pulls in when QI turns on and drops out when it turns off. The back contact of this relay controls the enlarger lamp circuit.

The interval of time required for the potential at the cathode-gate to rise to firing level depends on the charging rate of capacitor CI, which depends on the total resistance of resistor R2 and potentiometer RI. For a high resistance, the charging rate is slow and the timing interval is low.

The input resistance of the silicon controlled switch (SCS) at the cathode-gate is very high and turn-on current is low so there is very little loading on the RC timing circuit.

Diode D2 serves a dual purpose: it elevates

the potential required at the cathode-gate to fire the SCS and thereby enables a design with reasonable values for C1 and R1; and it assures reliable turn off when the cathode-gate is brought down to ground potential. In addition, selecting the optimum value for resistor R5, in series with D2 across the power supply, provides a quick and easy way to set the maximum timing interval. Initial adjustment of the maximum and minimum intervals is necessitated by component tolerances.

With switch S1 in focus, the enlarger lamp is on and the timing circuit is off. This permits adjustment of the enlarger for picture size, focus and lens aperture. In the reset position, the enlarger lamp is off, and the timer is off. This is the normal standby condition. For the TIME position, the enlarger lamp goes on and remains on for the interval of time determined by the setting of potentiometer R1.

When S1 is returned to its RESET position after a timing interval, C1 is discharged through R3, readying the circuit for the next timing interval.

Switch S2 enables the timer to be turned off when a darkroom session is over. However, no harm is done to the timer if it is left on.

1

are assembled on a printed circuit board as shown in Fig. 3. A foil pattern for this board is shown in Fig. 5. Use a glass-epoxy board to avoid leakage and possible mistiming due to moisture. This board is mounted on the top of the cabinet using appropriate spacers.

Depending on the relay used, its frame may be connected to the moving contact and therefore "hot" so the relay should be well insulated from the metal cabinet. Once all components have been assembled, wire the timer point-to-point in accordance with Fig. 1. Use dry-transfer lettering to mark the controls on the front panel.

Adjustment. When assembly and wiring are complete and checked, plug a table lamp using a 50- to 100-watt bulb into receptacle SO1. Turn on the lamp switch. Set S1 to FOCUS and S2 to ON. The lamp should light. Set S1 to RESET and the lamp should go off.

Now set potentiometer R1 to its maximum resistance position (should be maximum clockwise). Put S1 on time and observe that the lamp goes on. Using a stop watch or an electric clock with a sweep second hand, check the interval during which the lamp stays on. It

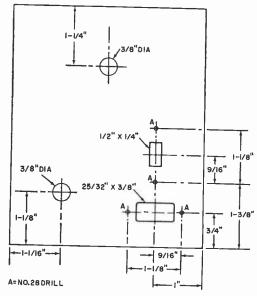


Fig. 2. To duplicate the front panel layout as used on the prototype, you can follow this quide.

should be somewhere between 65 and 70 seconds. If it is less than 65 seconds, unplug the timer and slightly decrease the value of R5. Then make the test again. Repeat the process if necessary.

(Continued on page 117)

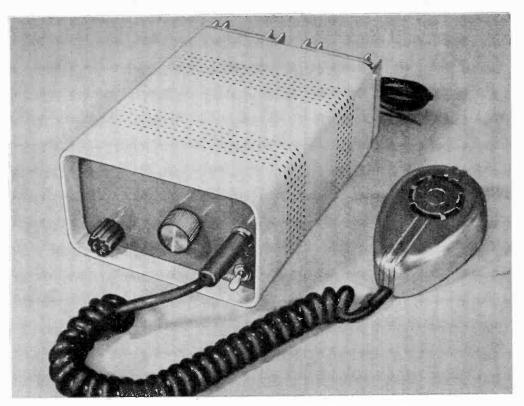
TO KI COIL 8 R2
DI R4 QI CI GND TERMINAL

TO KI COIL R5
TO RI TO PWR SUPPLY + R6

D4-D7

R3
R3
R3
SI

Fig. 3. The PC board containing the timing circuit is mounted on brackets to the top of the chassis. The power supply is mounted on its own board fastened to the top of the power transformer. Note that the relay frame is insulated from the metal chassis by a small piece of plastic sheet. In many relays, the armature is connected to the main frame.



TEN-WATT PA AMPLIFIER

INTEGRATED CIRCUIT SIMPLIFIES CONSTRUCTION

BY ED FRANCIS

WHEN A NEW electronic component is put on the market for experimenters, it is usually only after the commercial users have exploited it to the full. This is not so of the RCA CA3020 integrated-circuit audio power amplifier. From the moment of its introduction, the CA3020 was available to both commercial users and experimenters. It has been used in scores of applications.

The CA3020's success is due to a blend of small size, low cost, high reliability, ease of use, and a respectable ½-watt output. Even so, it is not powerful enough for public address audio amplifier applications. But, capitalizing on the IC's push-pull output configuration, it is possible to produce a PA system that develops a 10-watt output with the addition of a handful of components.

For maximum appeal, the PA system described here was designed for mobile/home use. All you need do after the project is assembled is connect it to a suitable d.c. power source, plug in mike and speakers, and you're in business.

About The Circuit. The CA3020 IC amplifier has an output stage consisting of two *npn* transistors in a push-pull configuration, requiring a 130-ohm centertapped load. Since the outputs (taken at pins 4 and 7 of *IC1* in Fig. 1) are 180° out of phase with each other, they are complementary-symmetry coupled to the bases of *pnp* power transistors *Q1* and *Q2* which are also connected in push-pull.

The input impedance of the 2N2869/2N301 transistors specified for Q1 and

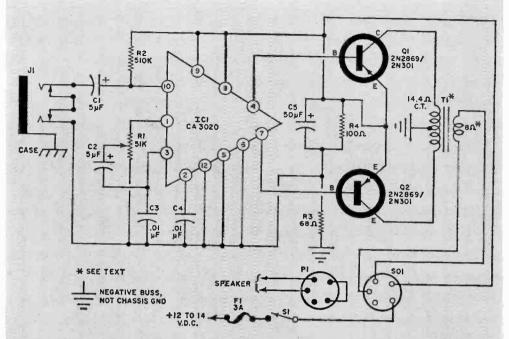


Fig. 1. Because IC1 cannot tolerate more than 9 volts d.c., connections to negative side of power source are actually 4-5 volts positive with respect to negative buss. Only connection made to the chassis is the J1 mounting bushing.

PARTS LIST

C1, C2-5-µF, 12-volt electrolytic capacitor
C3, C4-0.01-µF disc capacitor
C5-50-µF, 15-volt electrolytic capacitor
F1-Standard 3-ampere fuse (NOT slow blow)
IC1-RCA CA3020 1-watt audio integrated circuit amplifier
J1-Two-circuit control-transfer phone jack
P1-Five-pin molded plus (Amphenol 78-PCG5)
Q1, Q2-2N2869/2N301 power transistor
R1-5000-ohm audio taper potentiometer (IRC 60 137 6507)
R2-510,000-ohm, 5% resistor
R3-68-ohm resistor
K4-100-ohm resistor

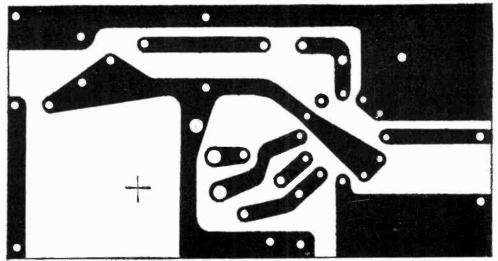
S1—S.p.s.t. switch (can be part of R1 or a separate slide or toggle switch)
S01—Five-pin socket (Amphenol 91-MPM5L)
T1—Output transformer (see text)
1—LMB Type CO-3 metal cabinet (Newark Electronics Corp. 91F1196)
2—Heat sinks (Olson Electronics X-951)
Misc.—Transistor sockets, bardware, mica insulators (2 each) for T0-3 case configuration; high-impedance (50.000-ohm) dynamic microphone with two circuit phone plug attached as described in text: 35%" x 134" printed circuit board; loudspeaker (see text); fuse bolder, control knob; three-lug terminal strip; rubber grommets; 1" metal spacers (6); 14/2" 6-32 machine screws and matching nuts and lockwashers; coded power cable; hookup wire; solder; etc.

Q2 serves as the necessary center-tapped load for IC1 and eliminates the need for a driver transformer. Transistors Q1 and Q2 feed output transformer T1. Since a complementary-symmetry configuration is also used here, no bias arrangement is needed for the power output stage.

The output stages operate on between 12 and 14 volts d.c., but IC1 cannot tolerate more than 9 volts d.c. Consequently, the voltage divider chain formed by R3 and R4 is incorporated into the circuit to provide a safe operating voltage for IC1. Because of this arrangement,

the common (usually grounded) pins of the IC are actually 4-5 volts above the negative buss reference; so it is necessary to use a two-circuit phone jack for J1 to prevent shorting R3. (Note that with this setup, chassis or case ground is not used as the ground reference for the input signal.)

Protection for the output transistors is built-in and the loudspeaker load must be plugged into the amplifier through the P1/S01 connectors before d.c. power can be applied. When P1 is unplugged, the positive input voltage line is broken.



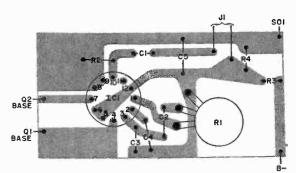
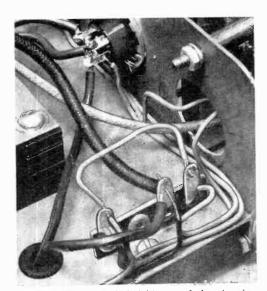


Fig. 2. When making printed circuit board, copy actual-size etching guide exactly as shown above. Then mount components on top of board (see drawing at left), making sure indexing tab of IC1 is located as shown.

No electrical connections are made to the amplifier cabinet, so the system can be used in either positive- or negativeground mobile electrical systems.

Construction. Apart from the fact that this is one project for which you will have to make your own printed circuit board (if you choose this method of assembly), construction is very easy. Except for the output transistors and transformer, the fuse holder, and the microphone and speaker/power connectors, all parts mount directly on the circuit board. If you prefer, you can assemble the circuit on a piece of perforated phenolic board, but a printed circuit board is recommended (see Fig. 2 for actual-size foil pattern guide and component layout).

Mount the components on the board in the following order: first the resistors, then the capacitors and volume control R1, and finally the integrated circuit IC1.



Terminal strip, mounted right rear of chassis, simplifies interconnections between power cable, output transformer T1, output transistors Q1 and Q2.

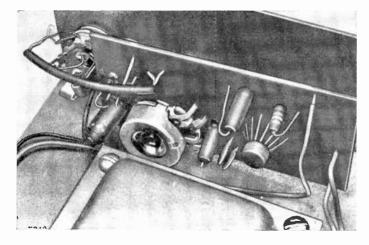


Fig. 3. Circuit board mounts to front of case via long threaded shaft of volume control. Use shielded cable to connect input jack to circuit board as shown.

When mounting *IC1*, leave the leads full length, carefully inserting them into the proper holes in the board. Then, heat sink each lead as you solder it to the foil. Temporarily set the circuit board assembly aside.

The next step is to homebrew an output transformer with specifications to match the circuit. To make this transformer, refer to the instructions provided in the sidebar on page 67. Then, mount the transformer inside the amplifier case in a position where it will not interfere with the circuit board or output transistor wiring.

Now mount the power switch (if it is not part of R1), fuse holder, and circuit board on the front of the case as shown in Fig. 3. Use two hex nuts and a control lockwasher to secure the extra-long threaded bushing of the potentiometer in place so that it protrudes from the front

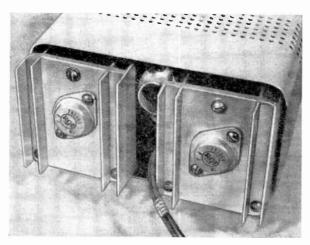
no more than $\%6^{\prime\prime}$, including the thickness of the hex nuts.

The heat sinks, as you get them, will not fit directly onto the rear apron of the case. Hence it is necessary to cut them down to size and file or grind the corners to fit the curvature of the case corners as shown in Fig. 4.

If the heat sinks are not pre-drilled to accommodate the TO-3 transistor and socket, machine them so that they do. Bear in mind, however, that care must be exercised in this operation to insure that, when mounted, the transistor leads or case do not short to the heat sinks. When you are satisfied, mount the transistors and their sockets on their respective heat sinks, placing a mica or teflon insulator and silicone paste between the transistor cases and heat sinks.

Space on the rear panel of the amplifier case is at a premium. Therefore,

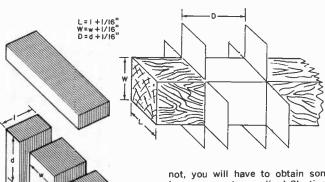
Fig. 4. Heat sinks for output transistors are slightly over size and must be machined to fit within chassis cowling; note rounded edges. They can be mounted to rear of case with spacers and hardware.



carefully select the location for SO1. Drill the hole for and mount this five-contact socket. Then solder 6" lengths of heavy-duty stranded hookup wire to the contacts of each transistor socket, and mount the transistor/heat-sink assemblies on the rear of the case. Use

1"-long metal spacers and 1\4"-long 6-32 hardware.

Now, referring back to Fig. 1, interconnect the circuit board, switch, fuse, power transformers and power transistors. For some of these connections, the job will be easier if you use a terminal



TRANSFORMER CONSTRUCTION

Since the output transformer needed for the PA amplifier is not a standard item, it must be homemade from a transformer that has been salvaged from an old radio or TV receiver. The only requirement is that the salvaged transformer have a core cross-sectional area (1 x w in upper left drawing) of approximately 0.7 sq in.

Disassemble the salvaged transformer as follows. First, remove the hardware securing the laminations together at the corners. Next, score through the weatherproofing shellac and remove one set of E and I laminations at a time, taking care not to bend them out of shape. If any of the laminations show signs of rust, clean them with steel wool or very fine emery cloth.

Now, referring to the drawing at the upper right, prepare a bobbin and winding handle. The winding handle can be a piece of scrap wood, while the bobbin material can be lightweight card stock (an index card, for example) or heavy-duty waxed paper, such as the type used for wrapping industrial tools or freezer wrap. Slide the bobbin onto the winding handle.

To insure that the center tap of the primary winding is exactly centered, fold a 10' length of #16 enameled wire exactly in half. (Check the original windings on the transformer. If the wire is the correct size, you can use it; if

not, you will have to obtain some #16 wire from your parts supplier.) Starting 4" from the fold, wind this doubled wire evenly up and down the length of the bobbin until there are exactly 26 double turns (52 single turns) on the bobbin.

Snip open the loop so that the folded end presents two independent leads. Now take one free lead from each end of the winding, and cut them so that they are just long enough to form a ½" pigtail when twisted together at the center of the bobbin. Scrape away the insulating enamel from each lead. Then twist together the two leads and one end of a 6" length of #16 or #14 stranded hookup wire, and solder the connection.

Cut the two remaining leads of the primary winding to the same length, scrape away ½" of the insulating enamel from each, and solder 6" lengths of #14 or #16 stranded hookup wire to each. These two leads should have a different color code from the center-tap lead to permit easy identification. Flatten all connections against the windings (do not allow the exposed connections to touch each other) and wrap a layer of electrical tape around the entire assembly.

Next, neatly wind the appropriate number of secondary turns (27 for 4-ohm output; 39 for 8-ohm output; 54 for 16-ohm output) over the tape-wrapped primary winding. Solder 6"-long #16 or #16-stranded hookup wire to the ends of these windings. Then tape the assembly as described above.

To assemble the transformer, slide the entire winding assembly off the winding handle and insert the crossbar of each E lamination into the windings core. Interleave these E laminations; do not insert them all in one direction. This done, slip an I lamination into place at the open ends of the E laminations. Replace the corner hardware. Then apply a coat of shellac to all exposed surfaces of the core laminations to provide a weather seal.

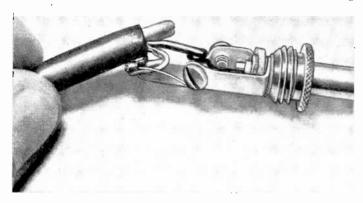


Fig. 5. When connecting plug to microphone cable, use only two shorter contacts. Do NOT allow cables to touch long one.

strip. Now wire SO1, and use a length of shielded cable to connect J1 to the circuit board. Then drill holes through the lower center of the rear panel and inner mounting base of the cabinet, insert a rubber grommet into each hole, and route the power cable through the holes, connecting it to the appropriate points in the circuit. (Note: the power cable should be coded so that the positive lead is clearly identifiable.) Assemble the amplifier case.

Disassemble the two-circuit phono plug that will be connected to the microphone cable. So that you do not make a mistake when connecting the cable, remove and discard the screw on the longest contact lug (there will be no connection made to this lug), and connect the conductors from the microphone cable to the two remaining lugs as shown in Fig. 5: This might appear to be an unconventional hookup, considering that the usual "common" contact is not being used. But this connection must be made exactly as described to prevent damaging the IC when power is applied and a microphone is plugged into J1.

Finally, wire the jumper across the appropriate contacts and the proper impedance speaker to its respective contacts of P1. Then slide the circuit assembly into the wrap-around case, and bolt the two pieces together.

How To Use. The input impedance of the CA3020 integrated circuit is 50,000 ohms. Although this is a fairly high impedance, the wiring of J1 described above precludes any possibility that hum will be introduced into the amplifier when the microphone is unplugged. With the

microphone plug withdrawn, the input of the amplifier is short circuited.

Now to use the amplifier, simply plug P1 into S01 and the microphone into J1. Being careful to observe the proper polarity, connect the 12-14-volt d.c. source to the power cable. Switch on S1 and, while talking into the microphone. slowly bring up the volume (rotate R1 clockwise) until the output sound level is perfect. Make sure, however, that the speaker is facing away from the microphone pickup or you will run into a feedback problem.

Only one precaution remains to be pointed out. The output transistors dissipate considerable heat; so it is necessary that you locate the PA amplifier where air will be allowed free circulation around the rear of the cabinet. If mounting space under the dashboard of your car, for example, is shallow, avoid mounting the PA amplifier there.



What's wrong with just a candle?



PERFECT TRIGGER SOURCE FOR THOSE DIGITAL PROJECTS

T'S ELEMENTARY! If you have built or are planning to build one of those digital instruments (timers, voltmeters, counters, etc.) that have become so popular since integrated circuits for digital applications were introduced, you'll need a pulse generator to check it out.

Pulse generators of many varieties are available—to be purchased or built—but most of them are too fancy and/or expensive for the needs of the ordinary experimenter. Here's one you can build for about \$12 (less, if your junk box is well stocked) and it meets all the usual requirements. It is compatible with both IC and discrete transistor circuits and it is adjustable in the three major parameters: pulse width, amplitude, and frequency. Specifications for the generator are given in the Table.

As a frequency source, this pulse generator uses a unijunction transistor, which operates over wide voltage ranges and oscillates from very low audio frequencies up to several hundred kilohertz. Its output trigger is easily adapted to conventional pulse design, and frequency is controlled by varying the applicable timing resistor or capacitor.

The pulse generator is portable, using a 9-volt transistor battery with a current drain of 9 milliamperes, no load.

Construction. The pulse generator, whose schematic is shown in Fig. 1, can

be built in an aluminum box about $6'' \times 4'' \times 3''$. The battery is secured to the rear panel by a metal clip or elastic band.

Assemble the electronic components on a $3'' \times 1' \frac{1}{2}''$ piece of perf board as shown in Fig. 2. (The transistor sockets are optional.) The controls and the output jacks are mounted on the front panel as shown in Figs. 2 and 3 and the front-panel photo. The perf board is mounted on stand-off spacers directly over the controls. The capacitors associated with S1 and S2 are mounted directly on the switches. Although component placement is not critical, lead lengths should be kept to a minimum.

The front panel can be lettered using any type of dry-transfer process, following the nomenclature shown in the photograph.

Operation. Although the Table of Specifications for the generator lists a maximum pulse width of 100 microseconds, the generator has a maximum duty cycle of about 75%. This means that, above a pulse width of 75 microseconds, the instrument "cheats" slightly and makes the width of every other pulse slightly narrower than it should be according to the setting of the controls. You will notice this discrepancy if you apply the generator output to an oscilloscope. It does not mean that the generator is not operating properly. If an attempt is made

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From The Leader HEATHKIT



NEW Heathkit Ultra-Deluxe "681" Color TV With AFT Power Channel Selection & Opt. RCA Hi-Lite Matrix Tube

The new Heathkit GR-681 is the world's most advanced Color TV with more built-in features than any other set on the market. Automatic Fine Tuning on all 83 channels ... power push button VHF channel selection, built-in cable-type remote control ... or you can add the optional GRA-681-6 Wireless Remote Control any time ... plus the built-in self-servicing aids that are standard on all Heathkit color TV's. Other features include high & low AC taps to insure that the picture transmitted exactly fits the "681" screen, automatic degaussing, 2-speed transistor UHF tuner, hi-fi sound output, two VHF antenna inputs, top quality American brand color tube with 2-year warranty. With optional new RCA Matrix picture tube that doubles the brightness. Model GR-681MX only \$535.00.

GRA-295-4, Mediterranean Cabinet shown \$124.95°

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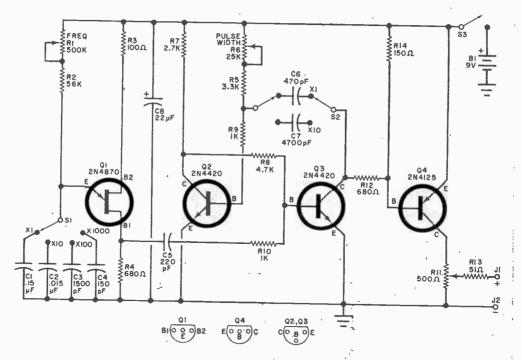


Fig. 1. The circuit consists of a variable-speed UJT oscillator driving a time-adjustable one-shot multivibrator. In this way, the narrow spikes from the UJT are converted into one of two selected output pulse widths.

PARTS LIST

B1-9-volt battery C1-0.15-µF capacitor C2-0.015-µF capacitor C3--1500-pF capacitor C4--150-pF capacitor C5--220-pF capacitor C6--170-pF capacitor CO-470-pr capacitor (7-4700-pr capacitor (8-22-pr, 15-volt electrolytic capacitor J1,J2-Banana tip jack 01-2N4870 unijunction transistor 02,03-2N4420 transistor 04-2N4125 transistor R1--500,000-ohm potentiometer R2---56.000-ohm

R3--100-ohm R4,R12-680-ohm R5--3300-ohm R7-2700-ohm All resistors R8-4700-ohm 1/2-watt R9.R10--1000-ohm R 13-51-ohm R 14-150-ohm R6-25,000-ohm linear taper potentiometer R11-500-ohm potentiometer S1-Single-pole, 4-position rotary switch S2-D.p.d.t. slide switch S3-S.p.s.t. switch on R1 Misc.-4" x 6" x 3" chassis, knobs (4), battery support, dry-transfer lettering, spaces (4), perf board, transistor sockets (4, optional), wire, solder, etc.

to increase the pulse width to more than 100 microseconds, the generator may start to divide the frequency. To insure stable operation, it is good practice to start with a narrow pulse width and then increase it as necessary.

Some times it is necessary to terminate the output in a low impedance. In this case, current drain will increase in proportion to the duty cycle, so it is a good idea to keep the pulse width as low as possible.

A 9-volt battery should give about 48

PULSE GENERATOR TECHNICAL SPECIFICATIONS

Frequency range: 10 Hz to 100 kHz Pulse width: 1 to 100 microseconds

Amplitude: 0 to 8 volts

Rise time: 10 nanoseconds at output terminals

20 nanoseconds with 3" coax lead

Fall time: 35 nanoseconds at output terminals

60 nanoseconds with 3' coax lead Output series resistance: 51 ohms

Power supply: 9 V at 9 mA, no load

Maximum pulse duty cycle: approximately 75%

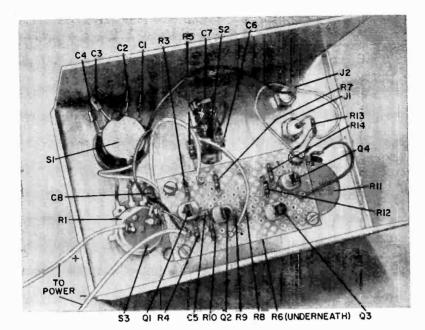


Fig. 2. Underside of chassis shows placement of board components and frontpanel mounted controls and terminals.

hours continuous operation in this pulse generator before performance is impaired. If you need more life, use six D cells. Needless to say, the generator should be turned off when not in use.

In checking the rise time of your pulse generator, remember that the rise time you measure can be no better than the rise time of the oscilloscope you are using. Also remember that, if too long a coaxial lead is used on the output, some degradation of the rise time may result due to capacitance in the cable.

If you want a negative going pulse, simply reverse the output lead. To obtain a negative going pulse which is

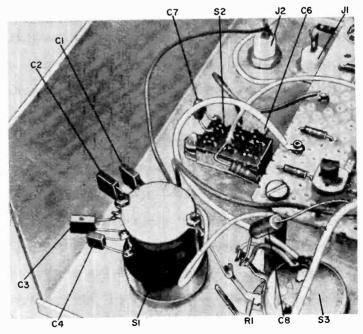
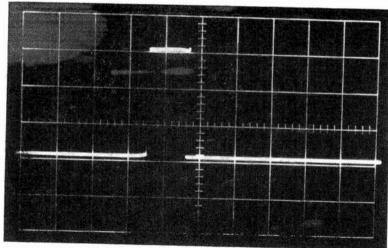


Fig. 3. Close-up view shows oscillator timing capacitors mounted on S1 and multivibrator capacitors mounted on S2. Perf board mounts on spacers.



Typical output pulse shows the extremely rapid rise and fall times on this pulse generator. The width of the pulse can be adjusted by the setting of switch S2.

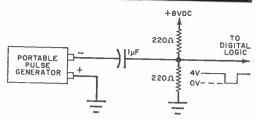


Fig. 4. If you want to drive 4- or 5-volt IC's with pulse generator, this external circuit is needed.

HOW IT WORKS

The basic pulse frequency is generated in the relaxation oscillator circuit containing unijunction transistor QI. Potentiometer RI provides the fine frequency adjustment while selector switch SI chooses the applicable multiplier capacitor (CI through C4). The selected capacitor charges up at a rate determined by RI and R2. When the capacitor charges up to about two-thirds of the supply voltage, the emitter of QI is forward biased and the capacitor discharges rapidly through the base-1 junction and R4. The resulting pulse is differentiated (narrowed) by C5 and applied to the base of Q3 through R10.

Transistors Q2 and Q3 form a monostable (one-shot) multivibrator. This circuit is in a stable state with Q2 saturated (collector voltage near ground) and Q3 cut off (collector voltage at supply voltage) until triggered by a pulse at the base of Q3. This causes Q3 to turn on, which forces the base of Q2 negative and turns it off. The selected timing capacitor (C6 or C7) then charges up at a rate depending on the total resistance of R5 and R6. Once the capacitor is charged up, the circuit reverts to its original condition, until the next pulse arrives from Q1. The result is a pulse at the collector of Q3 whose width is independent of the width of the trigger from Q1 but with a maximum width that must be slightly less than the time between input pulses. Also, the width cannot be reduced to less than that of the input trigger.

The negative (with respect to the supply voltage) pulse at the collector of Q3 saturates Q4 and its collector voltage rises, generating an output pulse across R11. The latter can be adjusted to give any desired amplitude. Resistor R13 minimizes overshoot of the output pulse and prevents damage to Q4 if the output is shorted.

referenced to 4 volts, reverse the leads and use the setup shown in Fig. 4. Since most integrated circuits operate on a 5-volt pulse and can be damaged if too high a pulse is applied to them, it is a good idea to mark the generator's dial at the 5-volt point.

The pulse generator is designed to drive all types of integrated circuits; however some forms require currentdrive capability while others supply current back to the generator. The generator will drive several RTL circuits because they draw current from the generator. On the other hand DTL and TTL types use the generator as a current sink and supply current to it. This means that the output resistance to ground must be kept low. Thus if several DTL or TTL circuits are to be driven and long coax cables are used, a buffer IC should be tied into the input of the IC circuit.

The Sterred Scerre by Charles Lincoln

INNOVATION IS THE NAME OF THE GAME

THAT NEW STEREO receiver with the low, low distortion, the extraordinary FM sensitivity, and all the rest—the one that brought you into the audio shop—how did it get there? What did it take to design and produce such a beautiful, complicated device?

To get the answer, Stereo Scene took a long look inside the process of creation and production of a representative first-grade American stereo unit, the Scott 342C stereo receiver. At the H. H. Scott plant in Maynard, Mass. we got a story that should prove to the audio buyer's satisfaction that the stereo receiver is one of the most remarkable consumer products turned out by any industry. Even a color television receiver, the only comparable domestic device, doesn't meet the exacting standards of performance we expect of a good stereo receiver.

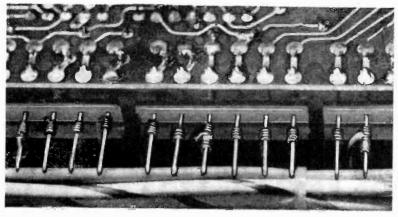
Stereo Scene is dividing this story into two installments. The first, published here, covers the conception and design of the unit. The second, coming next month, tells how the performance abilities of the 342C are safe-guarded in quantity production, a demanding and fascinating process.

For a product to run strong on today's very competitive high-fidelity market, it must embody at least one really good new idea, an idea involving a technical advance that will benefit the user materially. Of course, the idea must be practical and realizable within the cost framework prevailing in the industry. It is worth noting that the intense competitiveness of the American high-fidelity industry has provided many benefits for the audio buyer. He gets far more performance for his money than he did 6, 8, and 10 years ago. Consider, for instance, that a stereo receiver like the 342C. which sells for around \$300, includes an FM stereo tuner and two complete hi-fi amplifiers. A high-grade mono amplifier alone cost about the same amount 10 years ago, with no tuner! And today's unit will perform rings around the older one-the difference is almost ludicrous.

The Key—New Ideas. The conception of the 342C included several important new ideas. It was aimed, as we noted before, at the \$300 price bracket. That fact roughly established the power level and general configu-



November, 1969



Scott will probably be the first hi-fi manufacturer to use the wire-wrap technique. A special tool tightly wraps the wire around a square metal post. Perfect electrical contact is achieved and the connection is superior to ordinary soldering.

ration of the final stereo system design. The first of the new ideas was "modularization," the concentration of several vital sections of the receiver on standardized printed circuit units that plug into sockets on the main chassis. Using plug-ins was in large part a response by the engineering department to the growing severity of the servicing dilemma. The dilemma is that technical personnel capable of servicing complex high-fidelity units are getting scarcer all the time, while the units themselves, with dozens or scores of semiconductor devices tucked away in corners, get harder to service.

Now, if a 342C develops a rare fault, servicing is a breeze with plug-in modules. By simple test methods, he can isolate the trouble to one of the modules, and "fix" it simply by plugging in a new one. In a sense this takes us back to the "leap-frogging" method, when the radio-TV serviceman simply put a new tube into one socket after another until the trouble cleared.

The modularization idea made the most sense, the Scott engineers decided, if they could develop a set of standard modules that would be usable in a variety of stereo units. (The name "modules" implies this.) The functions finally chosen for modularization were the i.f. strip, the FM multiplex decoding section, the tone control section, the audio preamplifier, and the audio driver (two of the last for a stereo unit). Thus the design of the 342C was firmed up around those five standard units. The modules themselves, having been designed and tried out successfully in the 342C, are likely to turn up in other Scott stereo units.

The module story tells us quite a bit about the kind of engineering a hi-fi manufacturer needs in order to develop quality products—the engineering that goes into any receiver in the audio shop worth your unqualified admiration. The manufacturer must have a competent engineering team to originate both radio and audio design that will do a better job, at a practical cost. That demands a high level of creativity and long experience in turning creativity into practicality.

Integrated Circuits Make the Scene. Another important step in the conception of the 342C was the introduction of integrated



This is where it all ends—on the "burn-in" rack. These are model 342C's being individually tested before packaging. Details on the exhaustive test procedures will appear in the next "Scene" column.

circuits (IC's). As electronics enthusiasts know, these tiny, complex semiconductor devices contain whole sections of i.f. and audio amplifier circuits, the equivalent of scores of resistors, transistors, capacitors, etc. in one small unit. There will be wider use of IC's in stereo circuits, because, compared to circuits assembled from separate parts, IC's save space, eliminate dozens of soldered connections, speed production, and are potentially much more reliable.

Scott units already used IC's in the i.f. section. Now their use was extended to the preamplifier and FM multiplex sections. Since there were no available IC's of the right types for these circuits, Scott engineers worked closely with a semiconductor manufacturer to develop the necessary devices. As an example of what they came up with, the IC for the multiplex section is the equivalent of 40 transistors and 27 resistors!

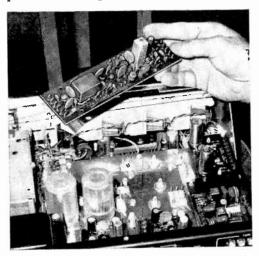
Also aimed at reliability and easier, faster manufacture was a decision to use the wire-wrap technique for many of the connections on the chassis outside the printed circuit modules. The Bell Telephone Company perfected the wire-wrap system to reduce the trouble they had with the millions of soldered connections in telephone central offices and exchange equipment. In the wirewrap system, a powered hand tool wraps several turns of the wire around a square connection post so tightly that the corners of the post bite into the wire and hold it permanently in place. You can't have a bad connection that looks good, as you can with a "cold" solder joint. The wire is either on the post for good or it doesn't stay at all. It's faster than soldering and takes less human operator skill; the skill is in the tool.

Another design decision was to use crystal i.f. filters to couple one stage to the next, instead of the tunable transformers that had always been used. Crystal filter coupling is coming into hi-fi receivers from satellite and space electronics. It has been used in one or two stereo receivers of a considerably higher price than the 342C. Crystal filters sharply reduce one of the most troublesome servicing problems: keeping the FM tuner in alignment (crystal filters never need retuning). They also have better pass-band characteristics, with broad, flat "tops" that pass the whole FM signal and a sharp cutoff for high selectivity.

Direct Output Coupling. The audio output circuit is representative of the type of amplifier design that was not possible before vacuum tubes were replaced by semiconductors. The output transistors are direct coupled to the speaker voice coil in each channel to eliminate the usual large-value

coupling capacitor—a source of much trouble. Careful balancing of the circuit keeps d.c. out of the speaker. The output amplifier is also free of distortion at low-power output levels that used to occur in many transistor amplifiers.

Something of a gimmick, but one that really helps, is "Perfectune," a circuit that turns on a special panel light when you are exactly tuned to the center channel of an FM station—a prerequisite for low distortion and noise. For many years, of course, tuning meters and electric eyes have told us where we were tuned on FM. But a circuit like Perfectune is more precise than the human eye focussed on a moving meter needle, in finding the exact center-channel point. Scott's engineers believe that, as ste-



Another innovation being introduced by Scott will be the modularization of certain circuits. The obvious advantage will be to pass along to the customer monetary savings in ease of replacement.

reo equipment becomes more complex, there will have to be more automation of control functions similar to the Perfectune. They see this as an important future trend in stereo design.

What we've described above are obviously just a few of the design decisions that are made in developing an advanced stereo receiver. But these items illustrate the kind of creative thinking that went into the stereo receiver on the audio shop shelf.

Model-Building Time. After the thinking, and the paper work that goes with it, there are four more long steps before a new receiver actually goes into production. First comes "breadboarding," with what look like jumbles of resistors, capacitors, IC's and transistors twisted together on the lab bench and connected to signal generators, os-

cilloscopes, and meters. Each circuit can be tried out this way with maximum ability to change things around. If a resistor, for example, turns out to be the wrong value, another can be connected in a few seconds.

Once the circuits have been proved out in this disciplined helter-skelter, an "engineering model" is built. This is the complete receiver with everything in its approximate place on a chassis of the desired size and shape. The model is put through extensive laboratory measurements and listening tests in a special reverberant room—as well as in rooms designed to be representative of actual use surroundings. The stylist studies the "model" carefully to develop a cabinet and front-panel layout.

When everybody is satisfied with the electrical and physical design, 25 or 30 "pilot models" are built of the final design. Scott considers this pilot model stage extremely important. Pilots are given to executives of the company, neighbors, and work-

ers in the plant for home trials of several days to a week. Does the user like the way the unit operates? Any problems? Each trial user makes a full, carefully detailed, subjective report

Since early in the design process, the manufacturing division has been working up a complete plan for producing the unit in quantity. This is an entirely separate problem from designing the unit, and an extremely complex one that takes a different set of specialized skills. By the time the pilot models have passed the shake-down, manufacturing is ready to go. That part of the 342C story will be told in next month's Stereo Scene.

Are you beginning to get an idea of what it took to put that stereo receiver, bright and shiny, on your audio dealer's shelf? Such products are innovative in the extreme, combining the knowledge and experience of a wide variety of engineers and designers in a very demanding field.

FIELD EFFECT TRANSISTOR PROJECT KIT

Although the field-effect transistor (FET) was developed long before the bipolar transistor, the FET has not been widely used until the last few years. In general terms, the FET exhibits many of the desirable characteristics of a triode vacuum tube in that it has a very high input impedance—up to 100,000 megohms in some cases, while still retaining some characteristics of the more common bipolar transistor (low operating voltage for example).

The FET differs from the more conventional bipolar transistor in the method of operation. Whereas a bipolar acts like a pair of back-to-back diodes, the FET is more like a high-speed, electrically variable resistor, whose instantaneous resistance is controlled by an electric field.

What held the FET back? Besides price, (it was high due to lack of demand), the big obstacle was a high noise factor. However, intensive research, improved materials, and advanced manufacturing techniques have led to a semiconductor product that holds great promise of becoming one of the stalwarts of the semi-conductor realm.

Things have gotten to the point where the ordinary electronics experimenter can now afford to play around with FET's. In line with this, Motorola is now introducing its HEK-2 HEP Field Effect Transistor Experimenter Kit (at most electronics suppliers for \$3.95). The package contains a pair of FET's—one for

d.c. to medium r.f. frequencies, the other for much higher frequencies—and a pair of conventional bipolar transistors to be used with the FET's. With the kit is a brochure entitled "Tips on Using FET's," which not only covers the theory of operation of the various types of FET's, discusses their various parameters, and tells how to test them, but also includes nine practical projects using the devices in the kit.

The projects include a three-semiconductor relay which consists of a basic circuit that can have one of three functions: timing from 5 to 50 seconds; detecting moisture (one drop of rain); or detecting the presence of light on a photo device. Other projects: a touch switch that operates a relay when a metal plate is contacted by a fingertip; a sound-activated relay controlled by any sharp sound such as the clap of the hands; a d.c. amplifier for extending the range of a scope, VTVM, etc.; a microphone or phono preamplifier; and a simple audio preamplifier with a high input impedance. For those interested in r.f. work: a broadband r.f. preamplifier with a range of 1 to 40 MHz and gain up to 100. And for the ham: a 14-dB gain, low-noise, 2-meter (150 MHz) preamplifier.

The HEK-2 kit is a good companion to the Motorola booklet "Field Effect Transistor Projects," priced at \$1.00.

the product gallery

REVIEWS AND COMMENTARY ON ELECTRONIC GEAR AND COMPONENTS

GLOBE PLOTTER (Megart Company)

THIS REVIEWER is always fascinated by new products that demonstrate originality -particularly if the simplicity of the product resolves a long-standing problem. This is the case with the Megart "Globe Plotter" (\$17.95). For years, hams and SWL's have tried all sorts of tricks to "visualize" the great circle route of a DX signal. Probably the most common solution is to hang a great circle (azimuthal equidistant projection) map on the wall. This works, but you'll find it impossible to buy a map less than three feet square and none that is suitable for just about any QTH. Also, the flat map leaves something to be desired-it doesn't "feel" quite right.

Megart is offering this cute globe that you arrange on a moveable pedestal to suit your own requirements. The globe measures 6" in diameter and is up-to-date. The printing is clear and there are ample country, city, and island identifications to suit almost

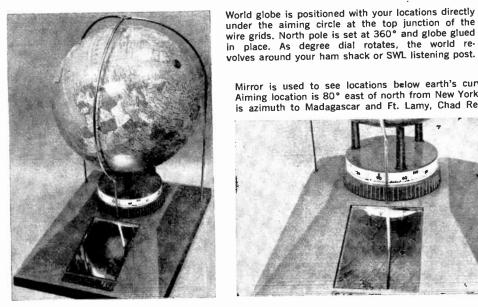
any DX'er.

You'll find on unboxing the Plotter that the globe is free of its pedestal. You then move the north pole around toward the guide wire facing the mirror and spot your QTH right under the aiming circle at the top of the frame work. Rotate the degree circle (holding the globe) so that it lines up with the north pole and, once you are satisfied that everything is in place, put 3 spots of glue on the pedestal uprights.

Now, as you rotate the degree circle, the world rotates around your QTH and the front wire guide tells the aiming azimuth toward a DX signal. Or, conversely, putting a DX station under the guide shows the path a signal follows to get to your QTH.

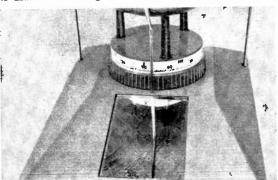
The value of the Plotter to hams with rotatable beams is obvious; perhaps not so obvious, however, is its value to SWL'ers. After using the Plotter for several weeks we realized that, when switching back and forth between two receiving antennas we were plotting the antenna lobe patterns. One antenna was great on 19 meters to Europe, but another antenna was better only 20° further east of north for stations in the Azores. Now as DX signals are heard and identified, we know which antenna to use and what countries "next door" to look for.

Circle No. 92 on Reoder Service Page 15



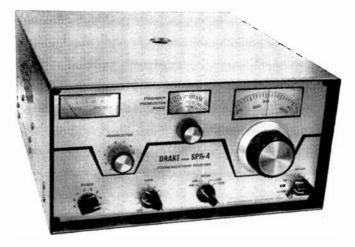
Mirror is used to see locations below earth's curvature. Aiming location is 80° east of north from New York which is azimuth to Madagascar and Ft. Lamy, Chad Republic.

in place. As degree dial rotates, the world revolves around your ham shack or SWL listening post.



November, 1969

CONTINUED



PROGRAMMABLE RECEIVER

(R. L. Drake Co. SPR-4)

F, AS CERTAIN advertisements say, you have a friend at such-and-such a bank, the shortwave listener is equally fortunate since he has an enthusiastic friend at R.L. Drake Co. The SWL got an inkling of this in 1966 when Drake introduced the SW-4 (now SW-4A) receiver. This was a product that tuned only the long-wave, medium-wave and shortwave broadcasting bands. The crystal-controlled front end permitted excellent linear dial calibration and the SWL could read out a frequency to better than one kilohertz.

The next logical step after the success of the SW-4 series was to go from tubes to solid-state, add more tuning ranges, add an internal speaker and tape recorder output, improve the selectivity. add optional noise limiting and crystal calibration, etc. It sounds like a big order. but Drake has fulfilled it in the brand new SPR (Solid-state Programmable Receiver)-4.

POPULAR ELECTRONICS had the opportunity to field-test one of the half dozen prototypes of the SPR-4. Matched against the SW-4A, there were some pretty outstanding differences. First, the dial mechanism has been simplified and made a little smoother (SW-4A and R-4B users, please note) and second, the selectivity is really meaningful. in fact, the SPR-4 has 3 selectivity positions that are automatically switched into the circuit as you move from straight AM (4.8 kHz at 6 dB down), to SSB (2.4 kHz), to CW (0.4 kHz). The skirts are much steeper and signals are easier to separate. As opposed to the 16-kHz-wide selectivity of the SW-4A (at 60 dB down), the SPR-4 is rated at 10 kHz

(-60 dB) a really meaningful improvement. Band selection and r.f. tracking on receivers similar to the SPR-4 have always constituted a problem. However, this has been solved by putting notations on the main Frequency Preselector Range dial using numerals and letters. The SPR-4 also has two thumb lever controls: one for an r.f. gain potentiometer, the other for selecting operation with a Noise Blanker or a Calibrator (both optional at extra cost) in the circuit. It is obvious that considerable thought has gone into making the SPR-4 more functionally useful for SWL's as well as laboratories, hams, etc.

In listening tests with the prototype, sensitivity and signal-to-noise ratio on all the short-wave bands were equal to a receiver costing 3 times as much. Selectivity was better. Medium-wave reception using the

CAPSULE CIRCUIT INFORMATION

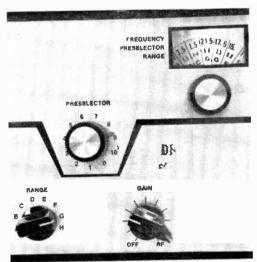
R.L. Drake SPR-4: All solid-state (dual-gate FET's) dual-conversion 24-band superhet, tuning 150-500 kHz and any of 23 ranges 500 kHz wide between 0.5 and 30.0 MHz. May be supplied to tune 0.5-1.0, 1.0-1.5, 6.0-6.5, 7.0-7.5, 9.5-10.0, 11.5-12.0, 15.0-15.5, 17.5-18.0, and 21.5-22.0 MHz. Three band-widths determined automatically by reception mode (AM, SSB, or CW). R.f. gain control. Built-in speaker. Built-in provisions for Notch Filter. Noise Blanker, 25-kHz crystal calibrator. Loop antenna input. May be used on d.c. with optional power cord. Manual fine adjustment for dial calibration. \$375.00.

loop proved interesting and locals could be nulled out without difficulty. Reception of SSB and CW was at least comparable to the Model R4-B-clean and easy.

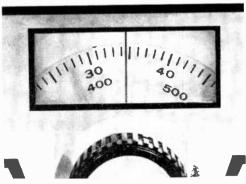
Our only arguments with the prototype (one of which was immediately corrected and won't be discussed here) concerned the S-meter sensitivity (200 microvolts for S9) and the placement of the headphone jack on the rear skirt (inconvenient to reach). Both of these may be modified before production models are delivered to the public.

All in all, the SPR-4 is a fantastic value and we'd say snap one up before the price goes up.

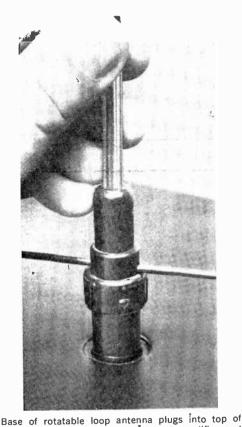
Circle No. 95 on Reader Service Page 15



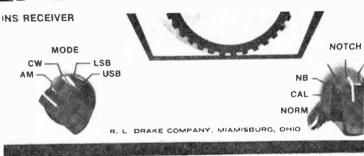
Don't get upset! We "phantomed out" the S-meter to emphasize the tie-in between the Range selector, Preselector and Range controls. Note the illuminated dial shows where each control is to be set.



the SPR-4. It is fed to an internal amplifier and is rotated to null out local medium-wave broad-casters. New dial design (left) has two rotating translucent discs—one calibrated in units and the other in hundreds. Add the figures shown to Range position and the frequency is 21.935 MHz.



Two thumb lever controls under the Gain and Notch knobs are for r.f. gain and either noise blanker or 25-kHz crystal calibrator use respectively.



ENGLISH LANGUAGE NEWS BROADCASTS FOR THE MONTH OF NOVEMBER

Prepared by ROGER LEGGE

	TIME-EST	STATION AND LOCATION FREQUENCIES (N	NORTH AMERICA FREQUENCIES (MHz)	TIME-EST	TO WESTERN NORTH AMERICA STATION AND LOCATION	AMERICA FREQUENCIES (MHz)
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		Montreal Canada	9.36, 11./1	8:00 a.m.	Stockholm, Sweden	15.315
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	7.47	Stockholm, Sweden	15.315	6:00 p.m.	Melbourne Australia	15.23, 17.625, 21.64
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	6:00 р.т.	Montreal, Canada	9.625 11.945 15.19	200	bonaire, Netn. Antilles	15.345
	7:00 p.m.	London, England	6.11 958 11 78	1,00	Jonannesburg, South Africa	9.705, 11.875, 15.22
		Moscow, U.S.S.R.	7.15 9 685 11.87	7:00 p:m.	London, England	6.11, 9.51, 9.58
		Peking, China	15.06 17.673		Madrid, Spain	6.14, 9.76
		Sofia, Bulgaria	9 70		Peking, China	15.095, 17.673, 21.735
	7:30 p.m.	Johannesburg, South Africa	9 705 11 875 15 22		Prague, Czechoslovakia	5.93, 7.345, 9.54, 9.63
		Kiev, USSR (Mon., Thu., Sat.)	7 28 9 685		Seoul, Korea	15.43
		Stockholm, Sweden	7.20, 3.003		Taipei, Taiwan	15.125, 15.345, 17.89
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,		Vatican City	6.145, 9.615, 11.785		Tirana Albania	11.705
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		Havana, Cuba	9 525		Havana, Cuba	9.525, 11.76
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ı		Cologne, Germany	6.075, 9.735		Cologie Comment	6.12, 9.72
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SOLID STATE

By LOU GARNER, Semiconductor Editor

LASER IN THIN FILM

BY USING a prism to couple a laser beam into a thin semiconductor film, scientists at the Bell Telephone Laboratories (BTL) have taken a promising step toward the integration of lasers and thin-film solid-state circuits. In the future, this technique could lead to the development of miniature laser amplifiers, light modulators, harmonic generators, and parametric oscillators, all of which would be useful in laser communication systems. In such miniature laser circuits, light beams would flow through thin transparent crystal layers, much as electricity flows in the copper wires of conventional circuits.

The basic idea of combining solid-state and laser technologies is not new; but before effective techniques could be developed, it was necessary to devise an efficient and practical means of putting laser beams into thin films.

Previous attempts concentrated on directing the beam through the film's edge. However, the ragged edges of semiconducting films tended to scatter the beam. In addition such films generally are much thinner than laser beams, often thousands of times thinner. Even if a beam could be focused down to the size of the film, the required precise alignment of the beam and film made this approach impractical.

In the BTL arrangement, the base of a

rutile prism is placed parallel to the film, but at a precisely controlled distance from it, to form a coupling gap, as illustrated diagrammatically in Fig. 1. The laser beam, entering the prism through its longest side, reflects from the base, according to the laws of optics. Contrary to convention, however, not all the laser energy is reflected.

Instead, a portion of the light waves "tunnels" out of the prism base, generating electric and magnetic fields which travel along the coupling gap at a speed set by the angle of incidence and the frequency of the laser beam. Under proper conditions, these fields transfer the laser energy into the film. Initial experiments have shown that over 50 per cent of the incident laser energy can be transferred into the film, while theoretical calculations predict a transfer efficiency as high as 80 per cent.

The light waves developed in the semiconductor film travel through it very much as do electromagnetic waves in waveguides ... that is, in predictable and distinct patterns of vibration, or modes. Acting somewhat as a dielectric waveguide, then, the film can support a number of different modes, each with its own characteristic speed of propagation. The different modes do not interfere with each other.

One major advantage of the prism-film coupler is that it can excite any selected one

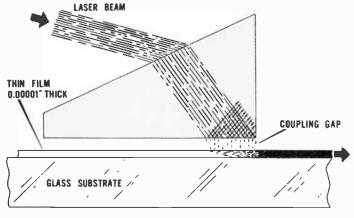


Fig. 1. Using a rutile prism placed parallel to thin film, laser beam can be made to enter film and travel through it like regular electromagnetic waves in standard waveguides.

November, 1969

of the possible modes simply by directing the laser beam at a proper angle, relative to the film.

The new technique is expected to be useful in a number of applications. Typically, it might be used to split light beams into different wavelengths to form separate channels of a laser communication system. Within the film, the beams could be modulated or amplified. Then the beams would leave the thin film via another prism-film coupler and travel to the destination point, probably along an underground "pipe line." At the destination, these beams again could be coupled into thin film circuits for signal processing.

Reader's Circuit. Intended for use in equipment employing electromechanical devices (motors or solenoids) as well as electronic circuitry, the dual-output power supply whose schematic is shown in Fig. 2 was submitted by reader Albert H. Reichel, Jr. (Rd. 2, Box 419, Monongahela, Pa. 15063). It can be used in control units, tape recorders or record players. Featuring a noncritical design and standard components, the project can be assembled in two or three evenings or on a weekend. The unit can supply up to 12 volts at 1 ampere through a single output stage, somewhat less if both outputs are used simultaneously.

All the items required for project assembly are readily available. Diodes D1 through D4 are 100-PIV, 1-ampere rectifiers similar to GE type A14A. Transistors Q1, Q2, and Q3 are general purpose pnp power units, type 2N554 or equivalent, while D5 is a 1-watt zener diode, with its voltage rating selected to meet individual circuit needs. Motorola types HEP-104 (9.1 volts) or HEP-105 (12 volts) are suitable. Capacitors C1, C2 and C3 are 15-volt electrolytics, while all resistors are half-watt types. A 12.6-volt, 1- to 2-ampere filament transform-

er serves as T1 (typically, Stancor type P-8130 or Triad type F-25X).

In common with most power supply circuits, neither parts placement nor wiring arrangement is critical, and virtually any construction technique may be used for assembly: perf board, chassis, breadboard, or etched wiring, as preferred. All d.c. polarities must be observed, of course, and the power transistors (Q1, Q2 and Q3) should be mounted on suitable insulated heat sinks. The completed unit can be installed in a metal enclosure or a commercial instrument cabinet, with either binding posts or plugtype jacks used as output connectors.

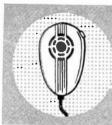
Generally speaking, regulated output "A" (Fig. 2) is used to power electronic circuitry, while output "B" is used to operate auxiliary motors or electromechanical devices (a turntable motor in a record player, for example). If only a single power source is needed, as in a radio receiver or audio amplifier, the isolated auxiliary output stage (Q3, R5, R6) may be omitted.

Manufacturer's Circuit. The touch-switch circuit illustrated in Fig. 3 was abstracted from "Tips On Using FET's," Bulletin No. HMA-33, published by Motorola Semiconductor Products, Inc. (Box 20924, Phoenix, Arizona 85034). Other projects in the publication include d.c., audio, r.f. and i.f. amplifiers, a sound-activated relay, and a general-purpose circuit which may be used as a timer, moisture detector, or light controlled relay. All the projects use low-cost components and are suitable for home assembly.

The touch switch is essentially a three-stage, direct-coupled amplifier. A FET, Q1, is used in the first stage to achieve an extremely high input impedance. In operation, the small control signal picked up on the metal touch plate changes Q1's instantaneous gate bias, supplying a d.c. signal to (Continued on page 105)

Q2 2N554 R3 1000 C3 DUTPUT 500 µF C2 + 1000yF R2 L5K 05 💇 220A Fر 500 5 0+ R6 3300 m 1000 DUTPUT LINE CORD AND PLUG

Fig. 2. Power supply to control electromechanical devices has two outputs of 12 volts at 1 ampere.



TWO HIS IONS

BY G. H. REESE, KCN6990

NEW PROGRAM IN NEBRASKA

NEBRASKA CB radio operators are being asked to serve as volunteer reporters in a new experimental program. This program will be known as PROJECT 20/20 and will research means to improve the handling of emergency and disaster messages. Also under study will be the prompt detection of such emergencies.

One of the significant weaknesses in emergency reporting systems—especially critical in rural areas—is the time lapse between the occurrence of an emergency or accident and the subsequent detection and notification of proper authorities. The usual information sources (state police patrols, county sheriffs, local police, the casual passersby, etc.) fail to insure rapid detection of even the most ordinary emergency or accident.

PROJECT 20/20 will coordinate a wide variety of reporters/travellers into a Volunteer Highway Surveillance System, known as NERS (Nebraska Emergency Reporting System).

Incorporated into NERS will be the reporting capabilities of several information sources:



Citizens Radio operators who are part of Nebraska Emergency Reporting System will display this symbol.

NERDASKA COLINTIES IN NERS

MEDIKAGIKA	000111120	111 11510
Adams	Howard	Otoe
Cass	Jefferson	Pawnee
Clay	Johnson	Richardson
Fillmore	Lancaster	Saline
Gage	Merrick	Sherman
Greeley	Nance	Thayer
Hail	Nemaha	Valley
Hamilton	Nuckolis	Webster

- State and county two-way radioequipped vehicles which are on the road a major portion of the time—Departmental Roads, Agricultural Extension Agents, etc.
- Vehicles that operate in rural areas on a regular schedule—RFD mailmen, milk tank trucks, bakery trucks, etc.

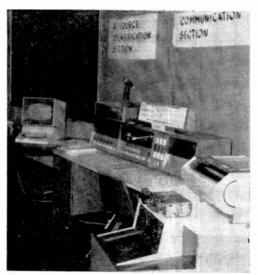
3. Citizens Radio volunteer groups.

The state of Nebraska has been divided into 8 Civil Defense emergency areas, and the 24 counties listed in the table are within two of these areas being utilized for Project 20/20. Each CB area will eventually have a centrally located Emergency Operating Center (EOC) equipped to communicate on designated emergency frequencies in each of the public service, industrial, business, military, governmental, and citizen's radio bands.

When notification of an emergency is received the location of the incident is plotted on a grid map, and the type of assistance required is determined. Selection of the assistance (police, fire, ambulance, doctor, etc.) nearest to the scene and best equipped to handle the emergency is made by a computer. The memory banks of the computer contain a central information source listing all of the services available within the Emergency Operating Sector. The computer readout is in the EOC, whose operator then telephones for the service required.

While some phases of this experimental system are not in full operation, the system will eventually be statewide and may serve as a model for national expansion.

Recently, Henry B. Kreer, REACT National Director, and Arthur Un rwood



Communications Center for Project 20/20 is a model of its type. Computer display provides information as to proper agency to notify in emergencies.

Manager of General Motors Research Laboratories, visited PROJECT 20/20 headquarters at Lincoln, Nebraska for a first-hand look at the test facility. They returned with a glowing report of the fine work being accomplished under the direction of Brig. General D. G. Pentermann, Deputy Adjutant General of the State of Nebraska

Citizens Radio users in the 24 Nebraska counties tied into the system are invited to enlist as NERS volunteers. This may be done as an individual, as an entire CB club, and as a REACT team. For information on how to become a NERS monitor, write to: Mr. John P. Nelson, Staff Assistant, PROJECT 20/20, 1300 Military Road, Lincoln, Nebraska 68508.

Travelers with citizen's two-way radio equipment will be able to call NERS monitors on channel 9 for assistance.

FCC Action. The FCC has requested comments on a proposed rule-making action that would amend the current Part 95 rules to exempt the use of CB radio to relay traffic information to licensed broadcast stations. This action is a result of a petition filed by National Capital REACT, Inc., Washington, D.C.

The change was requested after the FCC indicated that a proposed CB system of reporting traffic conditions to the AAA for subsequent broadcast was ruled in violation of the rule against transmitting messages for third parties. The FCC has now indicated that it intends to permit CB radio to be used for relaying traffic information for broadcast by licensed radio stations—and only that purpose! The public service aspects are apparently going to be recognized in this case and official approval may have been granted by the time you read this.

REACT National Headquarters has submitted its endorsement of the proposal along with the other leading CB organizations.

CB From A Wheelchair. Some of the most active Citizens Radio users are physically handicapped. Many find that CB radio opens a whole world that would otherwise be lost to them. For example, Ed Powell, Jr., KPK2106, of Chicago, Illinois, may be confined to a wheelchair, but his CB activity has given him the ability to reach out into the Chicago area and "meet" people while helping them.

Ed says that an interest in electronics led him into CB. After being licensed for 3 months, he joined a local REACT team and spent many hours monitoring Channel 9. During the record Chicago snow storm of 1967, he maintained contact with a number of local radio and TV stations, providing them with information about the emergency conditions as reported to him via CB radio.

Ed says, "When a young person is con-(Continued on page 102)

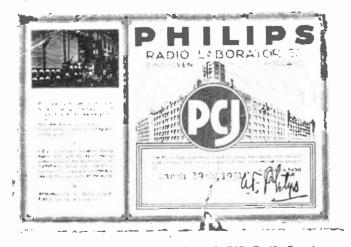
Ottawa Valley (Ontario, Canada) REACT team uses this trailer unit as a field base station.





PAR MORE than just casual interest has been shown in our discussion of old-time QSL cards (see April and July 1969 issues). Mr. Albert J. Sauerbier, WPE2NDA, of Washington, N. J., officer of the Newark News Radio Club, writes that he prizes as his oldest QSL's those from WSB, "The Voice of the South," Atlanta, Ga., dated August 13, 1923; WJAZ, Edgewater Beach Hotel, Chicago, Ill., dated two days earlier; WBAP, Fort Worth, Texas, from December 17, 1924.

and WOAI, San Antonio, Texas, with a date of January 15, 1926. On short-waves, Mr Sauerbier has a QSL from Berlin, Germany dated September 20, 1933; the League of Nations, Geneva, Switzerland, dated January 15, 1934, and one from EAQ, Madrid, Spain, dated March 18, 1934. Mr. Sauerbier says that the once-popular EKKO stamps came along at a later date and that, as a rule, you had to send a dime with your report in order to obtain a stamp. A person was considered





These interesting QSL cards from old-time broadcasters PCJ and LSX are from the collection of James Bruning, K2BZ, North Arlington, New Jersey.

81

NEW OPERATING SCHEDULES FOR INTERNATIONAL BROADCASTERS

Service Area Time	AUSTRIA	OSTERREIG	CHISCHER	RUNDFUNK, V	/IENNA	
Compose Comp				North America (General) 2300	-0400 6155
Europe (Roth)	Europe		6000			
Europe (East) 1000-1200 9770 Europe (Southeast) 1000-1200 9770 Europe (Southeast) 1300-1500 11,875 Europe (South) and 1300-1500 11,875 Europe (West) 1700-1200 Europe (West) 1700-120	Europe (North)		11 925			
Europe (Southeast)		1000-1200		South America		
North Africa						
Europe, North Africa, 0500-1300 7245		1300-1500	11,785			
Middle East		0700 0000	7045	South Africa		
Middle East				Middle Feet		
1300-1700 17,750 11,795 1400-1600 17,750 1700-2200 6155 Australasia 1000-1200 17,750 1700-2200 15,385 2000-2200 2100-2130 11,895 2200-4200 12,885 2200-4200 12,885 2200-4200 12,885 2200-4200 13,885 2200-4200 13,885 2200-4200 13,885 2200-4200 13,885 2200-4200 13,885 2200-4200 14,85 9640 English 1000-1200 17,780 12,885 2200-4200 14,85 9640 English 13,895 12	Middle East			middle Last		
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Company Comp		Time		Callsign	Frequenc	V Language
1030-1100	General Service					
Japan						
Japan						
1300-1400	Japan					
North America 1300-1400		0130-020	00			
Vietnam		0900-10	00			
Vietnam		1300.144	20			
Vietnam		1300-140	30			
Note	Vietnam	2230-230	00			
Southeast Asia						
Southeast Asia						
HLSA 970 English	Southeast Asia					
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HLKS		0830-090	00			
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Russia						
Russia						
Latin America	Russia					
North America	1 - A* A					
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A967.5				ASTING SERVI	- -	
9520 50 kW Arabic 0400-1500 Not specified 15,370 250 kW English 0400-0600 India, Pakistan 15,405 250 kW English 1600-1900 Europe 15,430 250 kW Arabic 0900-1100 Europe 21,685 250 kW Arabic 1300-1905 North Africa LIBERIA ELWA, MONROVIA Transmission Sunday Weekdays Frequency Languages Liberia and 0645-2245 0615-0815 4770 English, French, West Africa 1645-2300 African dialects Liberia 0615-1115 0610-0815 3225 Arabic, Liberian West and Cen- tral Africa 1200-1330 1200-1300 11,975 Hausa, Igala, 1600-2100 1645-2100 More, Yoruba, Twi Congo, E. Africa 1715-1900 1715-1900 15,155 French, Congo and Madagascar Middle East 1900-2100 1900-2100 15,170 Arabic						
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Transmission Sunday Weekdays Frequency Languages Liberia and 0645-2245 0615-0815 4770 English, French, West Africa 1645-2300 African dialects Liberia 0615-1115 0610-0815 3225 Arabic, Liberian 1805-2240 1805-2240 1805-2240 English, French West and Central Africa 1200-1300 11,950 English, French 1200-1330 1200-1300 11,975 Hausa, Igala, More, Yoruba, Twi 1600-2100 1645-2100 More, Yoruba, Twi Congo, E. Africa 1715-1900 1715-1900 15,155 French, Congo languages Middle East 1900-2100 1900-2100 15,170 Arabic	21,003	250 KW	Manic		1300-1905	North Africa
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Middle East 1900-2100 1900-2100 15,170 Arabic		1/13-1900	1/15-1	300	15,155	
	Middle East	1900-2100	1900-2	100	15,170	
	North Africa	2000-2200	2000-22	200		

twice blessed if he was able to get the stamps without the dime charge.

James Bruning, K2BZ, of North Arlington, N. J. listed as his three oldest QSL cards those from VRY, PCJ and LSX. Taking them one at a time, VRY was in Georgetown, British Guiana (now Guyar.a) and operated on 6840 kHz with 120 watts. They publicized their country as being the "Land of Many Waters and Eternal Summer." This card was dated October 12, 1930.

PCJ was owned by Philips Radio Laboratories in Eindhoven, Holland, and was the forerunner of Radio Nederland in Hilversum. Their QSL card listed the frequency as being 9.59 x 106 cycles/sec (9590 kHz). Power on March 19, 1931 (date on the card) was a mere 27 kW. They state on their card that PCJ commenced operations during March, 1927, with the epoch-making transmissions to the Dutch East Indies "12000 kilometers..." Announcements in those days were in Dutch, English, French, German, and Spanish.

LSX was operated by Transradio Internacional, Buenos Aires, Argentina, and operated with 20 kW on 10,350 kHz. The date on this card was December 29, 1930. Your Editor may be in error but we believe LSX is now operating in the coastal service rather than as a regular broadcaster.

Mr. Bruning was instrumental in the formation of the Pennsylvania State Police radio system. He designed and constructed WJL, a 500-watt station, that operated first on 1499 meters (about 200 kHz), later on 750 meters (400 kHz) and finally on 203 meters (about 1480 kHz). He has a reception report from a listener who tuned in on WJL on 1499 meters in 1924.

Transmissions from Greensburg, near Pittsburgh, were of unusual interest back in 1925. Harrisburg area headquarters personnel complained that they could not hear morning or afternoon broadcasts from WJL although early evening transmissions came in very well. On the other hand, morning transmissions were being heard hundreds of miles away to the west. Mr. Bruning states that this phenomenon, easily understood by any modern-day SWL, puzzled everyone including broadcast engineers from Westinghouse's KDKA.

Another station designed and built by Mr. Bruning was WBR, another of the police outlets, in Butler, Pa. Its output came from two fifty-watt tubes in parallel (not in push-pull). When first wired up and fired, it developed a nasty habit: first one tube, then the other, would turn brick red. After trying several remedies, none of which worked, a lead pencil wound with about 30 turns of fine wire with the two ends being connected to the two tube grids and the center point to the normal grid lead. The tubes stayed black and normal

power could be used. Modern-day hams would recognize this effect as "parasitic oscillations" with the wired pencil acting as a pair of grid chokes.

Bruning says, "I don't recall the call letters of the other two state police stations. However, my recollection is that Harrisburg Headquarters was WBAK and that their transmitter was donated by KDKA."

(Editor's note: My own recollection of tuning in the late 1930's and early 1940's was that Harrisburg Headquarters was then known as WPSP and that the frequency had been again shifted, possibly to 1674 kHz. Another reader, Ronald Shopinski, WPE3-DKA, Mt. Carmel, Pa., also informed us recently that the Pennsylvania State Police are now operating on 42,620 kHz.)

CURRENT REPORTS

The following is a resume of current reports. At time of compilation all reports were as accurate as possible, but stations change frequency and/or schedule with little or no advance notice. All times shown are Greenwich Mean Time (GMT) and the 24-hour system is used. Reports should be sent to Short-Wave Listening, P. O. Box 333, Cherry Hill, N. J. 08034, in time to reach Your Short-Wave Editor by the fifth of each month; be sure to include your WPE identification and the make and model number of your receiver.

Alaska—DX'ers who need this country should monitor 630 kHz for KYAK, Anchorage. Having obtained permission to operate with 25 kW, KYAK is now the most powerful station in Alaska and is expected to be an all-night station. It should be on the air now.

Andorra—Word has been received that R. Andorra has been operating in English since last May 30 on 5995 and 701 kHz at 2300-0300. Your Editor hasn't been able to pull them in; has anyone else?

Angola—CR6RZ, Emisora Oficial, Luanda, continues to be well heard on the West Coast from 0500 s/on with native instrument IS, anthem, and Portuguese language on 4820 kHz.

(Continued on page 96)



The well-appointed listening post of Daniel Trotto, WPE2QFZ, Little Falls, N.Y. features a Lafayette HA-225 and a Collins 75S-3B receiver. He also has an Allied 2682 receiver, an RCA TV, an RCA tape recorder and various pieces of CB equipment. Daniel now has 16 states and 40 countries verified.



INCENTIVE LICENSING. PHASE TWO

ON NOVEMBER 23, the second step of the Federal Communications Commission's revision of the amateur frequencies available to different classes of licensees goes into effect. Barring last-minute changes, the new frequency breakdown is as shown in the Table.

As the first anniversary of the FCC's Incentive Licensing regulations approaches, we have tried to evaluate its effects on ham radio by examining the FCC's monthly readout of the number of amateur licenses. In summary, the average number of amateur licenses in effect peaked at 259,000 in 1966, dropped to 256,300 in 1967, rose to 258,000 in 1968, and at writing is slightly above 258,000. Similarly, the number of Novices was 13,380 in 1966, dropped to 12,000 in 1967, rose to 13,540 in 1968, and to 15,080 in the first quarter of 1969. Also, the number of General class

licenses increased from 103,205 in September, 1965, to 108,595 at the end of 1968

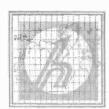
On the other hand, Technician, Conditional, and Advanced class licenses decreased 2800, 6165, and 2440, respectively. (It is interesting that, although new Advanced licenses had not been issued since 1952, this class showed less attrition than either Conditional or Technician.) To complete the picture, from the time the Extra class license was first made available in 1952 until the fall of 1967, the number of Extra class licenses increased at a fairly steady rate of approximately 25 a month, reaching a total of 4930 in October, 1967.

After the initial details of the Incentive regulations were released in late 1967, however, the number of Extra class licenses issued a month quickly increased by approximately

AMATEUR STATION OF THE MONTH



Donald Rubin, WA3JRA, 3919 Bancroft Rd., Baltimore, Md. 21215, likes DX chasing, contests, and ragchewing, in that order. He does all three with a Swan 500-C SSB/CW transceiver and a Mosley TA-33, triband beam. Already on his shack wall are Rag Chewers Club, Worked All Continents, and Worked All States certificates; but of the 115 countries worked so far, only 70 have QSL'ed therefore DX Century Club is 30 cards away. We are sending WA3JRA a 1-year Subscription for winning this month's Amateur Station Photo Contest. You can enter the contest by sending a clear photo (black and white preferably) of you at the controls of your station, plus some details about your amateur career to: Amateur Photo Contest. c/o Herb S. Brier, W9EGQ, Amateur Radio Editor, Box 678, Gary, Ind. 46401.

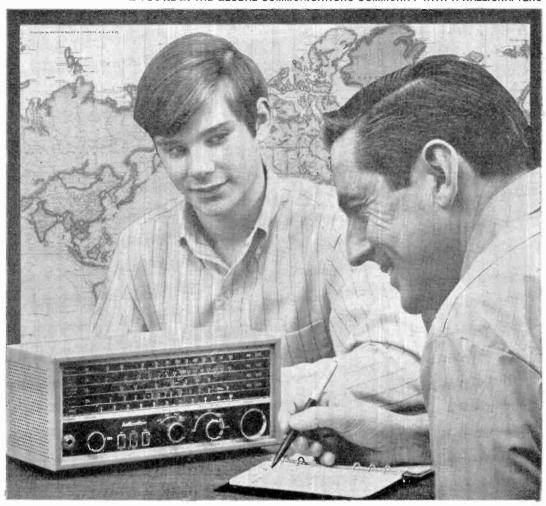


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700%, reaching a total of 85+1 in April, 1969. On the same date, the number of Advanced licenses had risen from its December, 1967, low of 36,691 to 46,215, while the number of General licenses was down to just under 100,000

Obviously, changing the Extra class license from a status symbol to a valuable possession has increased its popularity. In fact, so far in the Incentive Program, more Extra class than Novice class licenses have been issued! This fact casts doubt on the claims of some "authorities" that only geniuses can pass the Extra class test. At any rate, we hope that the increase in number of Novice, Advanced, and Extra class licenses encourages others to step up the amateur licensing ladder.

Antenna Tragedy. Via Bob Carter, W7INP, the Arizona Republic, and the WCARSentinel (Reno), we learn that three CB operators were electrocuted and three more were severely injured when the antenna tower they were erecting fell across a 12,000-volt power line. The tragedy emphasizes the importance of extreme care when installing an antenna anywhere near a power line. Even a 120-volt line can be lethal when an antenna mast falls into it. Furthermore, most utility poles carry power lines charged with far over 120 volts these days.

Amateur Radio in New Zealand. According to Break-In, journal of the New Zealand Association of Radio Transmitters, there were 4000 licensed amateurs in New Zealand at the end of 1968—an increase of six per cent in the year—and 10 per cent of the total participated in their December (midsummer) Field Day.

AMATEUR FREQUENCIES

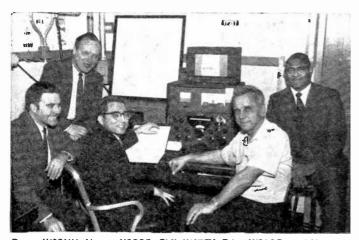
(November 23, 1969)

(
Class of License	Frequencies	
Novice	3.7-3.75, 7.15-7.2, 21.1-21.25, and 145-147 MHz (75 watts code only)	
Technician	50.25-54, 145-147*, and all amateur frequencies above 220 MHz	
General &	1.8-2**, 3.55-3.8, 3.9-4, 7.05-	
Conditional	7.2, 7.25-7.3, 14.05-14.2, 14 275-14.35, 21.05-21.25, 21.35- 21.45 MHz, and all higher ama- teur frequencies except 50- 50.25 MHz	
Advanced	1.8·2**, 3.55·3.8, 3.825·4, 7.05·7.3, 14.05·14.35, 21.05· 21.25, 21.275·21.45 MHz, and all higher amateur frequencies	
Extra	All amateur frequencies	

*At a later date, the 2-meter Technician assignment may be extended to cover the entire 144-148-MHz band.

**The 1.8-2-MHz band frequency and power assignments differ from state to state. Check with the nearest FCC office for your state.

Contests. The two big amateur contests of the fall operating season are upon us. The World-Wide DX Contest (CQ magazine) is scheduled for 0000 GMT, October 25 to 2400, GMT, October 26 (phone) and 0000 GMT, November 29 to 2400 GMT, November 30 (CW). These times translate to 7:00 p.m., EST, 6:00 p.m., CST, and 4:00 p.m., PST, Friday, until the same hour, Sunday, each weekend. Work as many DX stations as possible on the amateur bands up to 29.7 MHz; (Continued on page 92)



Donn, W2SNH, Henry, K3POD, Phil, W4TFX, Fritz, W3SQB, and Herman, WA3FVV, at W3FBE, the station of the Cadrerock (Md.) Amateur Radio Club at the Cadrerock Naval Ship Research and Development Center. (Official photo)

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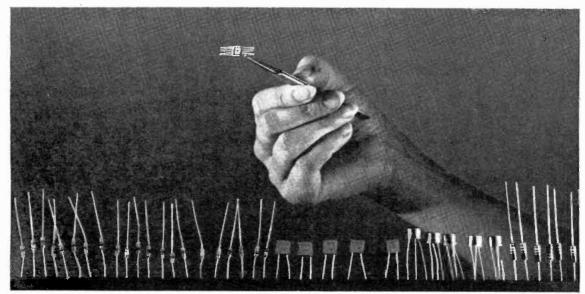
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Miniature Miracles of Today and Tomorrow

Already, as a result, a two-way radio can now be fitted inside a signet ring. A complete hearing aid can be worn entirely inside the ear. There is a new desk-top computer, no bigger than a typewriter yet capable of 166,000 operations per second. And it is almost possible to put the entire circuitry of a color television set inside a man's wristwatch case.

And this is only the beginning!

Soon kitchen computers may keep the house-wife's refrigerator stocked, her menus planned, and her calories counted.

Money may become obsolete. Instead you will

simply carry an electronic charge account card. Your employer will credit your account after each week's work and merchants will charge each of your purchases against it.

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What does all this mean to someone working in Electronics who never went beyond high school? It means the opportunity of a lifetime—if you take advantage of it.

It's true that the "chip" may make a lot of manual

skills no longer necessary.

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You can get the up-to-date training in electronics fundamentals that you need through a carefully

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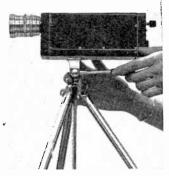
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AMATEUR RADIO

(Continued from page 86)

and exchange signal reports and DX "zone" numbers with each station worked. Check the Call Book for your zone number, or send a large return envelope with 18 cents postage to: WW DX Contest, % CO Magazine, 14 Vanderventer Ave., Port Washington, N.Y. 11050, for a free zone map and other contest information

The ARRL "SS" (Section Sweepstakes) Contest is possibly the world's most popular amateur contest, based on the number of participants. The 36th Annual Sweepstakes is scheduled for 2100 GMT, November 8 to 0300 GMT, November 9 (phone) and 2100 GMT, November 15 to 0300 GMT, November 16 (CW). Operate any 24 hours of the contest periods and work as many stations as possible in the ARRL "sections." Exchange message "preambles" as prescribed in the official contest rules. Send a stamped envelope to ARRL, 225 Main St., Newington, Conn. 06111, for the SS Contest package.



Jerry Dahl, WB4HUG, Charlotte, N.C., earned his Novice and General tickets through the code and theory classes of the Mecklenburg Amateur Radio Society. His equipment is a Heathkit SB-401 for transmitting and a Lafayette HA-500 for receiving.

News and Views

Landon Chapman, W4VTU, offers a free circular and folder to any would-be amateur in the United States who accompanies his request for information with a stamped, self-addressed number-10 return envelope. The address is Amateur Radio World, 201 Sunset St., Bristol, Tenn. 37520. Landon reports that he has helped 15 hams obtain their tickets. He also included an enthusiastic letter from a youngster of 74 who received his ham ticket at the age of 72 years. Unfortunately, neither of them mentioned the "youngster's" call letters . . .

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Beale E. Riddle, WA3MGA, 1808 Connecticut Ave., Washington, D.C. 20009, has a good antenna location and a fantastic landlord. His antennas are 200 feet above the ground on top of an apartment building on a hill, and the landlord says, "I don't care what you put on the roof—as long as it doesn't fall in the street." In three months as a Novice, Beale worked 33 states and four countries using a Knightkit T-60 transmitter to drive a base-loaded. 40-meter vertical antenna and a 15-meter dipole. He receives on a Hammarlund HQ-145. weeks after passing the General exam, WA3MGA passed the Advanced exam. If you need a volunteer examiner for your Novice or Technician exam or need Washington, Beale is your man . . . Gory W. Kent, WNØSZY, R.R. 1, Eddyville, Iowa, feeds



This photo is also the clever QSL card of Steve King, WN8CHK, Greenville, Ohio. He has 40 states with Heathkit DX 60B transmitter, HR-10 receiver.

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HAMARAMA

Annual Texoma Hamarama, November 14, 15, 16, at Lake Texoma State Lodge, Kingston, Okla. Fun for the entire family. Annual joint meeting with the Quarter Century Wireless Assoc. Plenty of camping and trailer locations. Registration orders to Texoma Hamarama, P.O. Box 246, Kingston, Okla. 73439. Write directly to the Lake Texoma Lodge, Kingston, Okla. 73439, for lodge or cabin accommodations.

a Hy-Gain 18-AVQ vertical antenna with a Heathkit DX-60B transmitter and receives on a Drake 2B assisted by a Heathkit HD-11 Q-Multiplier. Thirty-four states have been worked and confirmed, and Gary should now have received his General license.

Ronald Del Buono, WN2JUI, 415-41st St., Union City, N.J. 07087, uses a Heathkit DX-20 souped up to 60 watts to drive a 40-meter dipole. He didn't mention the model number, but his receiver is a Hallicrafters, and he worked five states and Canada the first few days he had his ticket. Ron listens to local 6-meter activity on a Heathkit Six'er and has 30 states and 30 countries confirmed as an SWL . . . Larry Skidmore, WN4LOE, 2855 Shady Oaks Dr., Titusville, Fla. 32780, seems to have almost two of everything. He transmits on an EICO 723 running 45 watts to either a 15-meter dipole or a Hy-Gain 18-AVQ vertical antenna. He receives on a Hallicrafters SX-99. Larry's first contact was with KP4AOD (Puerto Rico); he also has 32 states, and two foreign countries (Canada, and Cuba) in his log. Larry also set up a second



station at his school-the Titusville High Schoolusing a Hallicrafters HT-40 transmitter, SX-140 receiver, and a 15-meter dipole, which he keeps active during the school year . . . Robert Mahoney, WN8CXN, 26355 Normandy, Roseville, Mich. 48066, says that the biggest problem he faced in becomsays that the biggest problem he laced in second-ing an amateur was finding someone to be his volunteer examiner. In five months, Bob has worked 35 states using a Heathkit DX-20 trans-mitter, Hallicrafters S-76 receiver, and 40- and 15-meter dipoles. Next on his schedule is visiting the FCC for his General ticket . . . Maybe if enough qualified amateurs (over 21 and holding a General class license or higher) will follow Walt, WA3EOP's suggestion in August "News and Views" to indicate their willingness to act as volunteer examiners for prospective Novices and Technicians, we can alleviate the problem for other prospective amateurs.



Peter J. Harvey, WN2FWK, Delmar, N.Y., travelled the road from SWL, to CB, to Novice. He has worked thirty-nine states and 12 countries with an EICO 720 transmitter and a Hammarlund HQ-170 receiver.

David Mixerak, W?4???, P.O. Box 724, Burgaw, N.C., 23425, started as a Novice in 1966. He has worked 45 states and 52 countries since then using a Swan 350C, SSB/CW transceiver, Heathkit DX-60B transmitter, and National NC-109 receiver in conjunction with an 80-meter dipole and a Hy-Gain 14-AVQ vertical antenna. Dave is investigating the theoretical properties of semiconductors as a science project in High School . . . Congratulations to Adah Elliott, W9RTH, for receiving the Indiana Radio Club Council's 1969-70 Indiana Outstanding Radio Amateur Award; and to Phillip Haller, W9HPG, American Radio Relay League Central Division Director, for receiving the 1969 Illinois Amateur of the Year award. The awards were announced at the annual Indiana IRCC picnic in July and the Illinois Hamfesters Club picnic in August . . . Followup on the New England ham who forgot to give his call in the August column. He is Frank Arcivolo, WNIKIU, 16 Patten Rd., North Haven, Conn. 06473. Using his dad's (ex-W1NZM) HRO-7 receiver and a Heathkit DX-60B and a Windom antenna, Frank has now worked 26 states and has given his Dad the "itch" to get back on the air!

Will we see your "News and Views" and picture in an upcoming column? The first step is up to you. Thanks to you who have kept us on the mailing list to receive your club bulletins and announcements; if we do not receive your bulletin, we certainly would like to! The address is: Herb we certainly would like to: The addition B. Allos S. Brier, W9EGQ, Amateur Radio Editor, Popular Electronics, P. O. Box 678, Gary, Ind. 46401.
73, Herb, W9EGQ.

Name.

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CIRCLE NO. 39 ON READER SERVICE PAGE

SHORT-WAVE LISTENING

(Continued from page 83)

Antilles, Netherlands—Trans-World Radio. Bonaire, has moved it's English xmsn at 0225-0340 from 15,345 kHz to 11,775 kHz. A new, unlisted xmsn in Portuguese to South America has been found on 15,305 kHz at 2300-2330.

Australio—VNG. Lyndhurst, Victoria, a time station on 12,000 kHz, has been consistently strong in the west around 0600; time pips given each second except at the 59th second; ID on the hour.

Austria --Vienna is good on 9770 kHz from 2300 in German. New frequencies include 15,145 kHz from 9000 s/on to Central America and 15,200 kHz from 2120-2150 s/off with light music and multi-lingual annuts, including English, requesting reception reports.

Belgium—R. Belge. Brussels, has increased its xmsn's to N.A. and is now scheduled at 2205 on 11,715, 9615 and 9555 kHz and at 0050 on 11,715, 9615 and 6125 kHz.

Bolivia—CP90, R. Juan XXIII, San Inacio de Velasco, has moved from 4951 kHz to 4974 kHz where signals have decreased; it is heard only weakly at 0200 in Spanish talks with infrequent ID's.

Burundi—La Voix de la Revolution. Bujumbura, 6140 kHz. is heard at 0420-0450 with news in French, an ID, and a commentary in native language; two IS's are given after the news and commentary.

Canada—CKZU. Vancouver, noted on 6160 kHz relaying CBU at 1307-1344 with music, news, and a Peace River weather bulletin.

Chile—La Voz de Chile. Santiago, 9690 kHz, is heard in Spanish at 0100 with music program. It is easy to ID because of the extensive use of "Pomp And Circumstance" as an ID signal.

Colombia—HJZJ, one of the newer outlets of R.

Colombia—HJZJ, one of the newer outlets of R. Nacional de Colombia. 6030 kHz, has been noted lately with apparent test xmsn's until 0500 s/off dual to 4955 and 9635 kHz. English lessons are given at 2250. HJKC. Emisora Nuevo Mundo, Bogota, is audible on 4755 kHz after a Bolivian and a Brazilian on the same frequency close down; best listening time is after 0500 but there may be QRM from a Congo station.

Congo Republic—R. Brazzaville was noted on

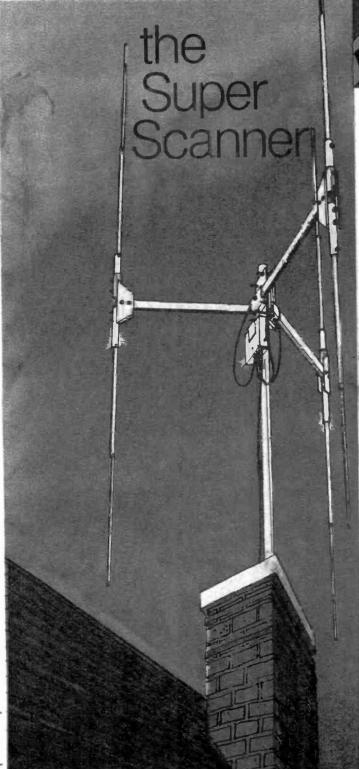
Congo Republic—R. Brazzaville was noted on 11.710 kHz with a special xmsn in French summarizing recent space events, including the moonwalk and landing of Apollo 11. This was from 2230-2258 s/off.

Costa Rica—TIRICA. La Voz de la Victor. San Jose, noted from 2345-0200 and 0430-0510 with pop music and mostly Spanish except for frequent anmt's in English offering free tourist information for prospective visitors. This is on 9615 kHz. TiQ. R. Casino, Puerto Limon, continues to transmit its English "East Coast Broadcast" from 0430-0600 s/off on 5954 kHz.

Ecuador—HCJB. Quito, has been heard on 11.880 kHz at 0015 in English and 15.300 kHz, dual to 17.880 kHz, in English to 2000, then into German. HCOS4. La Voz del Rio Carrizal. Calceta, 3570 kHz, was logged at 0500 with Spanish and music. HCCR1, R. Casa de la Cultura. Quito, last heard on 4930 kHz in 1960 when they stated they would soon have a 10-kW xmtr. is now on 4843 kHz with s/off generally around 0405; programming is mostly classical music.

El Salvador—R. Nacional, San Salvador, has moved from 6010 kHz to 5980 kHz where it is heard from 0255-0500 s/off with many speeches in Spanish. The dual channel of 9555 kHz is also heard, but much more weakly, also to 0500 s/off.

(Continued on page 100)



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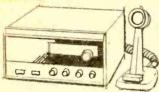
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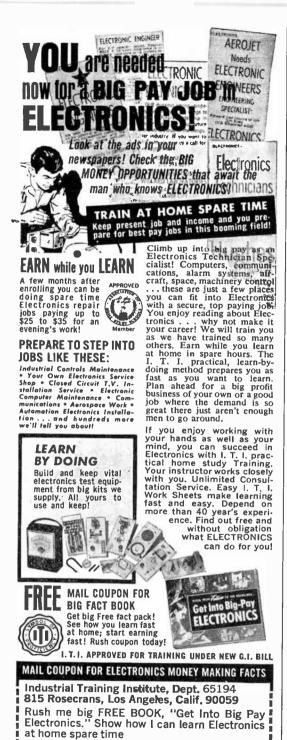
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CIRCLE NO. 21 ON READER SERVICE PAGE

_Zip

SHORT-WAVE LISTENING

(Continued from page 96)

Germany (East)—R. Berlin International was heard at 0100 on 11,890 kHz and at 1340 on 21,600 kHz, both with talks in English, and at 2000 with news in Spanish on 11,825 kHz. Another Spanish xmsn, to Europe, is aired on 11.840 kHz at 0220.

Germany (West)-R. Free Europe was monitored free of jamming on 11,825 kHz at 2000 in Russian. RIAS, 6005 kHz, has uninterrupted classical music at 0230-0315. Suddeutscher Rundfunk, Stuttgart, 6030 kHz, was logged at 0535-0555 with a nightclub-comedian type of program in German.

Guatemala-Radiodifusora Popol Vuh, 5070 kHz, is noted Friday 2230 to Saturday 0135 with a long talk to 0115, some songs, and an ID at 0130. The Popol Vuh (some sources spell it Buh) is said to be the sacred book of the Quiche race. Other facts are as yet undetermined but location may be Mazatenango. This is a rough one to copy and the only listening time is found to be as given above.

Guyana-One of our Ohio monitors, vacationing at Cape Breton Island, Canada, found ZFY, Demerara, Georgetown, with a good signal on 760 kHz at 1235-1305 with English and native language anmi's for American products.

SHORT-WAVE ABBREVIATIONS

anmt - Announcement GMT - Greenwich Mean Time ID-Identification IS--Interval Signal kHz--Kilohertz kW-Kilowarts

L.A,--Latin America N.A.--North America QRM--Interference R-Radio s/on-Sign-on xmsn-Transmission

Holland-R. Nederland, Hilversum, has returned to 11,730 kHz after a brief stay on 15,220 kHz with English to N.A., dual to 15.425 kHz, at 2125-2250. It's also heard at 2000-2030 in English to Europe.

Honduras-During the national crisis here, HRVC, Tegucigalpa, 4820 kHz, was monitored with its English "Master Control" program signing off at 0300 rather than at the usual s/off time of 0400.

Iran-Teheran's English xmsn at 2000-2030 is often good on 17,735 kHz. Two different IS's seem to be used, one being a stringed instrument, the other a clock ticking. Reports from the West Coast indicate a rapid dropoff in signal strength after s/on of the Russian xmsn which begins at 1730.

-Rome is on 15,410 kHz at 2230-0000 in Italian and at 0100-0120 in English, both to N.A.. and dual to 15.340 and 11,810 kHz. The 21,560 kHz outlet has Italian to Africa at 0435 and to Latin America at 1840 with both xmsn's being well heard.

Lebanon-Beirut, 15,285 kHz, has music at 0150, an ID and into Arabic at 0200, English at 0230-0300 to N.A. with news at 0230.

Papus—VLK3, Port Moresby, was heard on 9650 kHz from 0700-0715 s/off with a local interest program and recordings.

Peru—OBX7V, R. Tawantinsuyu, Cuzco,

moved back from the 49-meter band to 4910 kHz where it is heard with typical Andean music at 0200. OBX4U, R. America, Lima, is noted on 3240 kHz at 0340-0400 with L.A. music and Spanish anmt's.

Ryukyu Islands-Voice of United Nations Command, Deragawa, found on 14,460 kHz at 1453 with a male announcer in Korean,

South Africa-Low-frequency stations heard re-

City

Name.

Address_

SHORT-WAVE CONTRIBUTORS

John Duane (WPE1HBC), Hingham, Mass.
Joseph Breton (WPE1HKW), Methuen, Mass.
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Bob Raymond (WPE1HTD), Pittsheld, Mass.
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Bruce Roberts (WPE3KCB), Durbam, N. C.
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Carlton Blake (WPE5FTX), Fort Smith, Ark.
Richard Fortson (WPE5FDC), Dallas, Texas
Wallace Glavich (WPE5FTX), Fort Smith, Ark.
Richard Fortson (WPE5FDC), Dallas, Texas
Wallace Glavich (WPE5FDC), Dallas, Texas
Wallace Glavich (WPE5FDC), Salem, Ore.
Chuck Albertson (WPE7CWW), Salonhore.

Chris Kneibel (WPESIUB), Wooster, Ohio Bill Pelles (WPESIWU), North Olmsted, Ohio Robert Lewinski (WPESKAS), Wyandotte, Mich. David Cornell (WPESKA), Brockport, N. V. Wayne Shifflett (WPESKFA), Brackport, N. V. George Smith, Jr. (WPESKFA), Cand Rapids, Mich. Marty Himes (WPESKS), Chi'licothe, Ohio Tom Kennedy (WPESKS), Chi'licothe, Ohio Tom Kennedy (WPESKS), Indicago, Ill. Barry Aprison (WPE9ASH), Chicago, Ill. Barry Aprison (WPE9ASH), Chicago, Ill. Barry Aprison (WPE9ICS), Indianapolis, Ind. Bruce McCoy (WPE9IMV), La Porte, Ind. John Hedstrom (WPE9ICS), Indianapolis, Ind. Bruce McCoy (WPE9IMV), Park Forest, Ill. A. R. Niblack (WPE9IJVX), Park Forest, Ill. A. R. Niblack (WPE9IJVX), Park Forest, Ill. A. R. Niblack (WPE9ASE), Pueblo, Colo. C. Vernon Hyson (WPE9CWK), Somerville, N. J. Jim Randles (WPE9CWK), Cunningham, Kansas Fred Mc Cormack (WPE9DSU), Salina, Kansas Jack Perolo (PT2PEIC), Sao Paulo, Brazil Paul Cau (VE3PE2OZ), St. Catharines, Ont. Art Strandquist (TE4PEFG), Winnipeg, Man. Bruce Archinal, Dover, Ohio Lloyd Berman, Van Nuys, Calif. Kim Bodkar, Detroit, Mich. Carl Borden, Piedmont, Calif.
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cently include R. RSA, Johannesburg, 4895 kHz, at 0500 with English news, and Springbok Radio, Paradys, on 3952 kHz at 0450 in Afrikaans and on 3997 kHz at 0430-0450 with a morning get-up show with exercises and piano music.

Sweden—R. Sweden, Stockholm, is scheduled to eastern N.A. on 11.950 kHz at 0000 and 0130 in Spanish, 0030 and 0200 in English and at 0100 in Swedish, and on 15.315 kHz at 1100 and 1400 in English, 1130 and 1430 in Swedish and at 1500 in French: to western N.A. on 11.705 kHz at 0300 French, 0330 English and 0400 Swedish, and on 15.315 kHz at 1600 English and 1630 Swedish. Also noted well are English xmsn's at 1400-1430 and 1600-1630 on 21.585 kHz.

Turkey—R. Ankara is heard well in English at 2200-2230 and 0200-0230 on 15.160 kHz and at 1415-1445 on 17.820 kHz. Programs contain news, comment, music, and weather warnings for Turkish navy boats.

USA-NSS, U. S. Navy station, Annapolis, Md., has time signals from 2255-2300 with an ID in Morse a few seconds later other xmsn's are in all Morse. This one, often used by Your Editor for time correction, is on 5870 kHz.

USSR—R. Kiev has English to Europe at 1930-2000 on 15.150. 11.700 and 9740 kHz, and to N.A. on Monday. Thursday and Saturday at 0030-0100 on 12.010. 11.960, 11.730 and 9600 kHz and at 0430-0500 on 12,010, 11.730, 11.700 and 9600 kHz. Kiev's 17.900 kHz outlet has also been found in English at 0030 dual to 17.760. 15.260 and 11.700 kHz. Minsk, Byelorussia, 15,150 kHz, has a R. Moscow relay to N.A. at 2300 in English. Mrs. Stepanova is reportedly now issuing verification cards marked "via Minsk." R. Yerevan, 17.880 kHz, is heard at 0318 with nusic, at 0324 with news in Armenian and in English at 0328 closedown. USSR Home Service from Irkutsk has an unusual IS at 1200, then into Russian, on 11,905 kHz.





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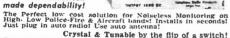
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TWO-WAY REACTIONS

(Continued from page 80)

fined to a wheelchair it is difficult to learn things that he would normally learn through experience. Thus CB is a great educational tool. It can give you many valuable experiences without leaving home. Just listening to the radio is informative because all kinds of people are on the air. You get exposed to people you could meet no other way."



Members of Gateway REACT (Waveland, Miss.) as well as many others, provided emergency communications plus general assistance in the wake of hurricane Camille. They brought much-needed supplies of all types in addition to their 3500-watt generator.

CURRENT NEWS

San Diego, Calif. . . . Southwestern REACT recently participated in a unique communications activity in connection with the annual Powder Puff Derby, a transcontinental air race. Using a callsign assigned by the FCC for the event, they relayed the takeoff of each entrant from San Diego's Lindberg Field to E. A. Johnston, KMX8506, who is also a member of the West Coast Amateur Radio Association. Mr. Johnston then initiated the first in a series of six relays between San Diego and Washington, D.C., to keep all points advised of times of departure for the 92 women flyers. This is an outstanding example of cooperation between Citizens Radio users and ham radio operators.

Atlanta, Ga. . . . Summertime is usually a quiet period for most organizations-but not CB clubs and REACT teams. For example, Capital City CB Communications, Inc., started out the summer with a retreat and training exercise at Fort Yargo State Park. At 1:00 A.M., on July 3rd, they were called out to search for a missing woman. Then after finding her in the woods when daylight came, they rested up to prepare for 4th of July parade communications. On July 14th, 30 club members enrolled in a 30-hour Red Cross First Aid course. On July 25th, 4 members, supported through CB

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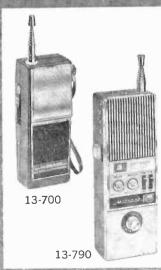
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communications with other members, won the WQXI Great River Raft Race. August 22-28 was the occasion of the National Convention of the American Legion in Atlanta. CB radio was used to coordinate communications for the Legion parade. Lasting over 6 hours, it was the longest ever in Atlanta history, Another retreat was held at Fort Yargo State Park on September 13 and 14. In addition to all this activity, the Capital City Club maintained its REACT monitor 24 hours daily to assist the many additional vacation-bound motorists in the Atlanta area.

An interesting incident took place in the course of this monitoring function last summer, According to CCCBC President, Terry Simonds, a stranded motorist radioed for assistance to "Capital City REACT". The Atlanta monitor answered the call in the customary manner. When the motorist responded with an unfamiliar location, another REACT monitor offered to be of assistance. This was Capital City REACT in Sacramento, California. They were the object of the motorist's call. He was on a freeway in Sacramento, California and the signal "skipped" to Atlanta Georgia.

DON'T FORGET

If your club is interested in becoming a REACT team, write us for information. Remember? "Two-Way REACTions!"

We are interested in receiving information from everyone involved in Citizens Two-Way Radio. Send us all the news about your activities so that we can include it in this column. Be sure to include photos! Remember, we are also interested in a celebrity list of Citizens Radio users. Send direct to "Two-Way REACTions," c/o REACT National Headquarters, 205 West Wacker Drive, Chicago, III. 60606.

Among summertime activities are the many jamborees, coffee breaks, barbeques and picnics held by clubs. There are so many of these that we simply cannot list them all. It is sufficient to say that if one wished to drive 100 miles or more in any direction on a summer weekend, he could attend an event sponsored by a citizens radio group. Our publication deadline is so far in advance that it is not too soon to send in information for next summer's jamboree calendar. We actually need four months notice to be sure of publication.

While we give emphasis to the public service aspects of CB activities, we must recognize that there are purely social groups. They use a common interest in two-way radio as a means of drawing people together purely for fellowship. Sometimes these social clubs also participate in civic events, but that is secondary. These CB clubs often do much to encourage the positive use of Citizens Radio among their members, In addition, they are prepared when called upon to serve others with emergency communications. Among the groups of this type who regularly send us their bulletins are: Tri-County Citizens Radio League of Akron, Ohio; Waukesha County CB Club, Inc., Oconomowoc, Wisconsin; Lycoming CB Radio Club, Inc., Montoursville, Pa.; 7-11 CB Radio Club, Spokane, Washington; Sociable 5 Watts, Inc., Enon Valley, Pa.; Holiday Citizens Banders, Baltimore, Maryland; Cereal City Citizens Band Radio Club, Battle Creek, Michigan; Chicago Citizens Radio League, Chicago, Illinois; Citizens Radio Association of Lake County, Waukegan, Ill.; Five-Eleven Radio Club, Inc., Pittsburgh, Pa.; York CB Assistance Club, York, Pa. and Calgary General Radio Club, Calgary, Alberta, Canada.

SOLID STATE

(Continued from page 78)

Q2's base and causing a corresponding change in its collector current. This, in turn, controls Q3's base bias. Transistors Q2 and Q3 together form a complementary d.c. amplifier. A conventional electromagnetic relay (K1) acts as Q3's collector load, and as the control switch for an external device.

The active devices used in the project were selected from Motorola's popular HEP line. A 15-volt electrolytic capacitor is used for CI, while all resistors are half-watt units. The relay should be a moderately sensitive type having a 12-volt coil, such as the Calectro type D1-967. The power source (B1) may be either a line-operated d.c. supply, a 12-volt lantern battery, or eight series-connected penlight or flashlight cells.

Well suited to either perf board or etched wiring construction, the touch switch can be assembled in just a few hours time. Although neither layout nor lead dress are overly critical, Q1's gate connection leads should

be kept as short and direct as possible. The touch plate can be any small piece of metal, with neither exact size nor shape important as far as circuit operation is concerned . . . use your imagination!

In practice, the relay serves to actuate an external electrical device whenever the touch plate is touched lightly by the user. The external device, operated by a separate power source, may be a safelight in a darkroom, a solenoid, a small motor, an electric light, or an alarm signal, depending on individual user needs. The relay's contacts are used as a simple switch to turn the external device "on" or "off," as needed.

New Devices. RCA Electronic Components (Harrison, New Jersey 07029) is continuing to expand its line of linear integrated circuits. Among the devices which should be of special interest to experimenters and hobbyists are a series of three high-gain wideband amplifiers with low current and voltage requirements, making them ideal for battery powered projects. All three are assembled in 12-lead TO-5 packages. The CA3021, with a 2.4-MHz bandwidth, requires only 4 mW, for operation, yet can supply 56-dB gain

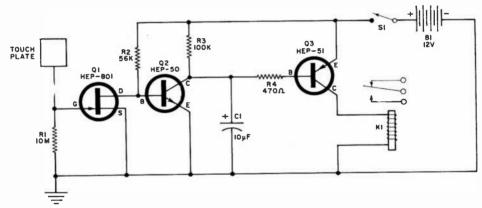


Fig. 3. A FET is used in the first stage of touch switch amplifier to achieve high input impedance.



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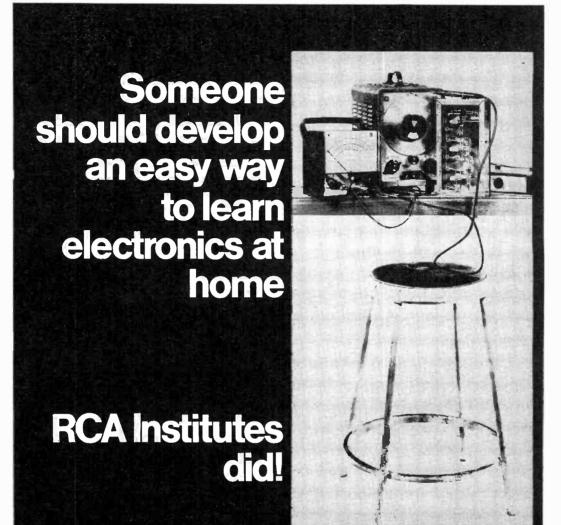
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Automation Electronics. Gets you ready to be an Automation Electronics Technician; Manufacturer's Representative; Industrial Electronics Technician.

Automatic Controls. Prepares you to be an Automatic Controls Electronics Technician; Industrial Laboratory Technician; Maintenance Technician; Field Engineer.

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using a 6-volt supply. Furnishing up to 57dB voltage gain, the CA3022 has a 7.5-MHz bandwidth and needs only 12.5 mW, at 6 volts. Maximum bandwidth is supplied by the CA3023, with a 16-MHz rating and a power requirement of but 35 mW, at 6 volts when furnishing 53-dB gain.

A pair of complementary transistors suitable for 100-watt audio amplifiers, a new 90-watt plastic encapsulated unit, and two sets of 50-ampere complementary devices are among the new silicon power transistors recently introduced by Motorola Semiconductor Products, Inc. (P.O. Box 20912,

Phoenix, Arizona 85036).

Packaged in modified TO-3 cases, pnp type MJ4502 and npn type MJ802, can each replace up to three parallel transistors in audio amplifier circuits supplying up to 100-watt output. The units have d.c. current gains of 25 to 100 at collector currents of 7.5 amperes and a power dissipation rating of 200 watts at 25°C.



Complementary pnp and npn transistors have d.c. current gains up to 100 for total output of 100 W.

A new monolithic amplifier-detector silicon integrated circuit designed primarily for control system applications has been introduced by the Sprague Electric Company (115 Northeast Cutoff, Worcester, Mass. 01606). Identified as the ULN-2301M, the new device is a linear differential amplifier with a relatively high input impedance (70,000 ohms), 37-dB gain, and a frequency response of 150 kHz. Assembled in a modified 8-pin dual in-line molded plastic package, the unit features an internal zener diode voltage regulator and is intended as a control amplifier/driver for medium power SCR's.

Amperex Electronic Corporation (Slatersville, Rhode Island 02876) has announced the development of a complete AM radio receiver IC on a single monolithic silicon chip. Designated the TAD100, the device is offered in a 14-lead, dual-in-line package, and contains all the active components of a complete AM receiver, including the oscillator, mixer, i.f. amplifier, detector, AGC, audio preamp and driver stages. The unit requires external tuning elements and con-

POPULAR ELECTRONICS

trols. The performance of a TAD100 receiver is comparable to that of a similar set made from discrete components. Its AM sensitivity is 50 µvolts/meter for 100 mW, audio output, its a.g.c. range is 65 dB for an audio output change of 10 dB, and its total harmonic distortion is typically 2%.

The TAA300 is another new Amperex IC device. A 1-watt audio amplifier, the unit is packaged in a 10-lead TO-5 case and has a sensitivity of 10 mV for full output, including 20 dB feedback. Its input impedance is over 10,000 ohms, while its total harmonic distortion is only 10% and its noise figure is typically less than 6 dB from 30 Hz to 15 kHz. The device is intended for use in battery-powered portable equipment such as phonograph amplifiers, radio and TV audio systems, walkie-talkie audio and speech amplifiers, tape-cassette systems and miniature hi-fi sets.

Transitips. Increasingly popular among solid-state equipment manufacturers, directcoupled circuits require servicing techniques which differ somewhat from those used to troubleshoot conventional capacitortransformer-coupled designs. With each stage dependent upon all preceding stages for correct biasing, a minor defect near the input can result in a later stage being biased to cut-off or saturation. The familiar signal tracing technique may not isolate the defect, for an (a.c.) signal may appear to be normal until stage biases exceed rather broad tolerances.

One of the most effective test techniques is a stage-by-stage check of d.c. bias levels, starting at the last (output) stage and working back towards the input. Naturally, power supply voltages should be checked for normal levels before individual stages are tested.

If the input (base) bias is normal in a specific stage, but its output d.c. levels are far off, further checks of individual components in the stage generally will isolate the defective part. If all fixed components test "good," suspect the active device.

There are exceptions to this general rule, however. A leaky or shorted input circuit can affect the preceeding stage. Here, one must consider the following stage's input circuit as an active part of the stage under

test.

Finally, never overlook the possibility of a defect in an early stage causing parts damage in later stages. An initial minor defect in a preamp, for example, may cause a later driver stage to overload, damaging, say, a load resistor by excessive heating. Once a defective component has been located and replaced, then, recheck the entire circuit. -Lou

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MICROWAVES

(Continued from page 46)

tered between the transmitter and receiver antennas. Make one of the antennas fixed; leave the other movable. Turn on the power at the transmitter and observe that the receiver readout indicates upscale. This proves that r.f. energy radiates from the transmitter, hits, and is reflected back from, the metal sheets, and produces a reading on the receiver.

Slowly slide the movable reflector toward the antennas. You will soon reach a point where the received power drops to a null and then rises again. This is the half-wave point where half of the transmitted r.f. is being supplied to the receiver out of phase with the signal from the fixed reflector. Carefully note and record the position of the movable reflector as indicated by the ruler. Keep sliding the movable reflector in toward the antennas, recording the exact points where nulls occur.

Calculate the centimeters or inches difference between each ruler indicated, add them up and divide by the number of nulls measured. This will give the average half wavelength. Double the value to obtain the full wavelength.

For the band being used here, full wavelength for 2300 MHz is 13.04 cm or 5.13 in.; and for 2450 MHz, 12.24 cm or 4.81 in. If your wavelength is not within these limits, bend tuning inductor L1 to change the frequency. Retune the receiver to the transmitter either with C2 (if the frequency difference is not too great) or by bending L1 in the receiver.

Microwave Experiments. Once your microwave system is operating on frequency, you can experiment with various types of antennas and reflectors. For instance, with the transmitter and receiver facing each other a couple of feet apart, note that placing a metal sheet above the transmission path will greatly boost the signal. This is similar to the way the ionosphere reflects low-frequency r.f. signals.

Placing a pair of metal sheets near either antenna so that the antenna is at

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the apex of a right angle also boosts the signal. This is a basic horn antenna. It can be used at either the transmitter or the receiver, or both. The signal strength can be vastly increased by various antenna configurations, even though the transmitter power remains unchanged.

To demonstrate this further, make a dipole as shown in Fig. 9. Connect the center feed to the thin coaxial cable to the transmitter antenna coupling capacitor C2 and the shield to the ground plane. Disconnect the regular antenna. Now you can add directors and/or reflectors to the basic dipole to experiment with beam arrays. A couple of things will become apparent as you experiment: First, as gain goes up, directivity increases; second, to double the range (distance), you must have four times the power. The latter can be proved by using the potentiometer on the transmitter to vary the power. A milliammeter can be inserted in the 12volt line to measure the transistor current.

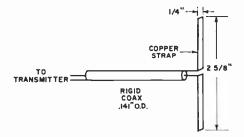


Fig. 9. This simple dipole can be used to perform experiments with various antenna arrays. These can include beams, horns, or parabolic systems. Remove the old antenna and ground plane and install this.

In experimenting with various horns and arrays, remember that, to get about 20 dB of gain, the antenna aperture should be at least six wavelengths square (or 36 square wavelengths) at the aperture.

The transmitter can be audio modulated by conventional techniques and such a system makes a very interesting Science Fair project. Microwave signals can be "bounced" around corners and over other intervening objects, by means of simple metal reflectors. When a hand is placed in the beam path, the signal level changes, showing that the microwave r.f. is actually carrying the signal.

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AIR IONS

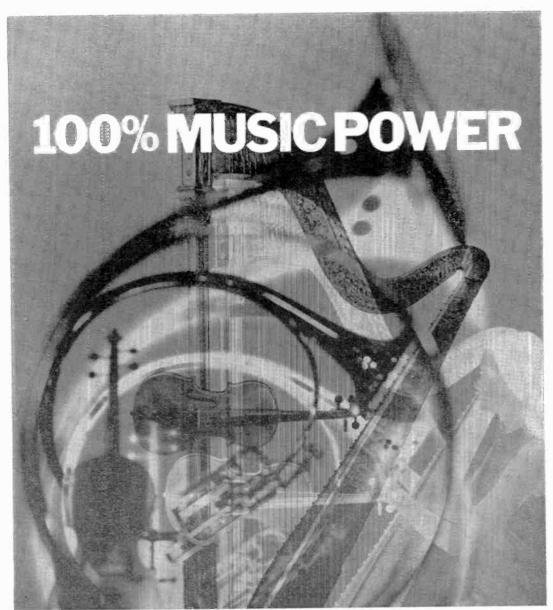
(Continued from page 30)

large ions in the air is not controlled. Work has been done in this field in an attempt to improve conditions in closed quarters (such as submarines).

European scientists were among the first to research the relationship between ions and human health. A great emphasis in clinical work with ions has developed within the past few years. The effects on tissue and cell life of animals and humans are increasingly popular areas of research. The field is attracting the physicist, biologist, climatologist, and medical researcher. In this country additional work has been done by the heating and air conditioning interests in an effort to determine the factors that are involved in human comfort.

The Air We Breathe. One area of research which has received considerable attention is the effect of ionization on hay fever, asthma and other respiratory troubles. Some of the early work in this field was done by Dr. Kornblueh in 1953. As a result of his early experiments he found that symptoms of hay fever and asthma were relieved when patients were exposed to negative air ionization in a closed room. However, the symptoms returned when the patients left the room. In later experiments he established a hay fever clinic at Northeastern Hospital in Philadelphia. Under controlled conditions 63% of the patients received complete or partial relief under negative ionization. The symptoms returned when patients returned to their normal environment. Positive ionization gave no relief and in some cases increased the distress of the patient.

Research has shown that negative ionization has an unusual effect on the bronchial tubes and trachea. These portions of the respiratory system are lined with tiny hairlike filaments called cilia which maintain a wave-like motion. With this motion they keep the air passages free of dust and pollen. When an excess of positive ions was maintained in the intake air, the wave-like motion decreased by several hundred beats per minute.



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When the tissue was exposed to an excess of negative ions the motion was accelerated above normal values. It was also found that under positive ionization the tissues were much more susceptible to bruise or damage.

Other studies have shown that negative ions have been used successfully to treat hypertension, slow-healing wounds, and to reduce the pain in severe burn cases.

The details of the picture puzzle are being filled in piece by piece but several important details are still missing. As one researcher put it, we do not know why ionized air has a physiological effect on living things but it does have an effect on all of us.

Another point on which most researchers agree is that negative air ions do not cure anything. During the time that they are being applied they apparently set up a favorable condition or perhaps supply something that is missing in the individual being treated. When the supply of negative ions is removed the condition returns. However, this does not diminish their value. They should be very useful in pointing the way to what is defective or lacking in the patient. Just the relief of symptoms has its value.

One item that should not be overlooked is the fact that most negative ions are formed around an oxygen molecule and most positive ions are formed around carbon dioxide or sometimes nitrogen. The charge may be important only as the nail that holds the package together.

To solve the riddle of the ion and to apply the knowledge gained could be a breakthrough as important as the moon landing, at least to those who will get relief from a serious ailment. The fascinating thought is that some reader may just do it in his back-room laboratory.

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In the next issue, Popular Electronics will publish the most exciting construction project in its history—a safe, but fully operational LASER!

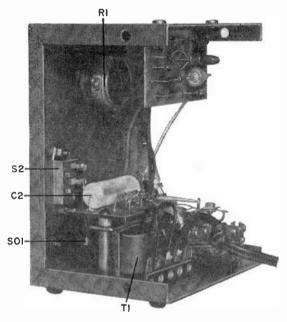
Subsequent issues will show how to use this laser for secret communications and making holograms.

Don't miss this most important issue of 1969-70.

DARKROOM TIMER

(Continued from page 54)

With the maximum timing interval properly set, turn potentiometer R1 to its minimum resistance position. Using the stop watch or clock, check the minimum timing interval. It should be somewhere between 4 and 5 seconds. If it is longer, slightly decrease the value of R2 and check the interval again: (Selecting the optimum value for R2 has no significant effect on the maximum timing interval.)



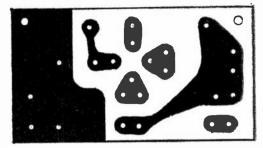
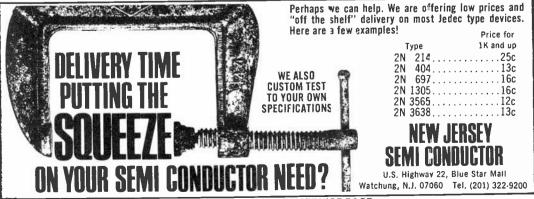


Fig. 5. Actual-size printed circuit foil pattern can be duplicated. Use a glass-epoxy board to reduce moisture absorption and possible mistiming.

Calibration. The timer may be calibrated with the same stop watch or clock used in making the adjustments. Mark the 10-second intervals first. As each position of the pointer knob on R1 is accurately located, make a tiny prick point on the cabinet front panel. Do the same for the 5-second intervals. Then remove the knob and apply decals or dry transfer at these points. Make certain the knob is replaced in exactly the same position in which it was taken off.

Fig. 4. This view shows the placement of the outlet socket S01, input power control switch S2, and timing potentiometer R1. Make sure that the cover can be installed without touching any components.





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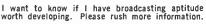
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CIRCLE NO. 10 ON READER SERVICE PAGE

PRODUCTS

(Continued from page 24)

ture automatically rotates to each of the receiver's channels every half second. When one channel is active, scanning stops at that channel until the message is complete. Then the scan engages again until stopped by another active channel. The scan circuit, however, can be disengaged if desired simply by flipping a switch located on the front panel. The Mark 6 is also available as an excellent receiver without the "Scan-O-Matic" feature.

Circle No. 89 on Reader Service Page 15

VHF/FM MONITOR RECEIVER KITS

Two new monitor receiver kits are now available from the *Heath Company* for coverage of the 108-136 MHz (Model GR-98) and 154-174

MHz (Model GR-88) bands. The GR-98 is for aircraft signal monitoring, while the GR-88 is for monitoring VHF marine channels, weather bureau broadcasts, fire and police calls, commercial radio, etc. Both receivers feature six-to-one vernier tuning controls to make it easy to tune closely-spaced stations, and



built-in whip antennas. Selectivity and sensitivity in both cases are 40 kHz and 1.5- μ V, respectively. If continuous monitoring of only one station is desired in either model, separate plug-in crystals are available. The receivers have adjustable noise-operated squelch, operate on six C cells, and feature a carrying handle that converts into a tilt stand.

Circle No. 90 on Reader Service Page 15

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dio spectrum with both channels driven simultaneously into an 8-ohm load. Harmonic distortion is less than 0.5%, and i.m. distortion is less than 0.1% at the rated output, decreasing as power is reduced. The amplifier is said to operate with reliability and stability into such reactive loads as electrostatic speaker systems. The control section has complete facilities, yet is simple to operate with a basic, two-knob control action. A front-panel stereo headphone jack, always live, can be used with speakers either live or off.

Circle No. 91 on Reader Service Page 15
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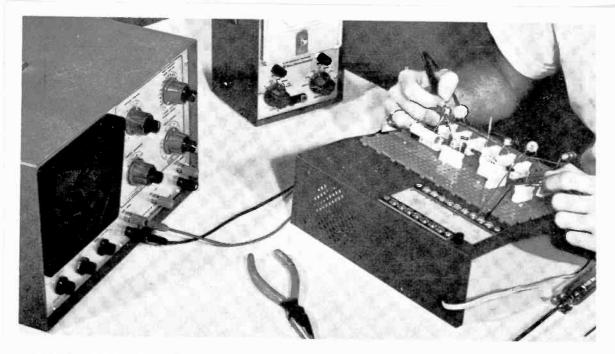
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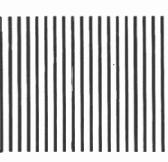
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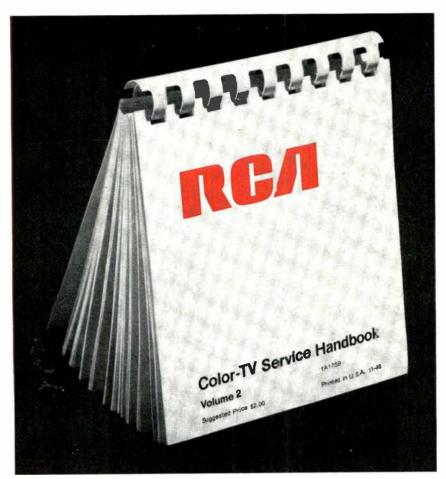
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