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COVER PHOTO: Frank A. D. Andrea and his daughter inspect one of his early "Neutrodyne" kits in front of a present-day Andrea television set. (Kodachrome by Nick Desmond)

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RADID ERGINEER ABG RELWORK "4 years ago, I was a bookkeeper with a hand-to-mouth salary. Now I am a Radio Engineer with key station of the ABC network."—NORMAN H. WARD, Ridgefield Park, New Jersey.

St TO SIO WEEK IN SPARE TIME "While learning, made \$5 to \$10 a week in spare time. Now have a spare time shop in my home and earn as high as \$25 a week"-LEANDER ARNOLD, Pontiac, Michigan.

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GETS FIRST JOB THROUGH N.R.I. "My first job, with KDLR, was ob-tained for me by your Graduate Serv-ice Dept. Am now Chief Engineer, Police Radio Station WQOX."-T. S. NORTON, Hamilton, Ohio.

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RADIO-TELEVISION SHOPS NEED SHOWMANSHIP

THE past few weeks have been spent by this editor on an extensive field trip in order to gather ideas from radio and television dealers and technicians and to discuss problems of sales and maintenance. We have had the opportunity to "spot check" cities and towns, both large and small.

Eleven states were included in order to make visits in both TV and non-TV areas. We wanted to find out what radio and television dealers were doing to increase sales, what promotion they used, what test equipment was doing to speed the repair of sets, and what we, as publishers, could do to help make their businesses more profitable.

It didn't take many stops to discover that, in comparison to other merchandisers, radio and television shops were far behind in the matter of display of sets and components and that the average service shop presented an unsightly mess to its customers.

The trend in automobile dealer service departments has been toward neatness, cleanliness, and "showmanship." It was only a short time back that such service to automobiles was rendered in space cluttered with old parts, with greasy floors, inadequate lighting, and without specialized equipment.

Today the aggressive auto dealer takes great pride in the appearance and efficiency of his repair department. He knows, from experience, that customers—especially women—won't hesitate to bring in their cars to a wellequipped shop where cleanliness, efficiency, and courtesy prevail.

And so it is in the case of radio or television service shops. We technicians can learn much of specific value by observing methods and techniques employed successfully by other businessmen that cater to the same public.

We can, and economically, tidy up our service benches and rearrange our test equipment to be both workable and presentable. We can clean up the messy (and dangerous) wiring of the line supply on the bench. We can do a bit of painting and arrange our parts cartons to good display advantage. These things come under the heading of "showmanship."

The wearing of a neat shop coat does much to impress the customers. They cost but little and can pay real dividends to the technician.

Many of the dealers we visited actually featured and pointed with pride to their service departments. These, in most instances, were situated at the front of the store by the window for all to see. We were impressed by this method of "showmanship" and can readily see its advantages.

In other establishments, however, the service bench was back in a dark corner and was cluttered with old chassis and dust laden cabinets from as far back as 1930. Visible to customers, these shops certainly couldn't expect to make a good impression with those who had occasion to bring their sets in for repair.

It is particularly important that service technicians who own and use modern, up-to-date service equipment bring it out into the open and give their customers a true appreciation of the "professional" character of the work involved in servicing radio and television receivers.

Good commercial test equipment represents a large investment on the part of the shop owner. Failure to exploit its "showmanship" value is simply missing a good and no-investment possibility.

Conversely, if the test equipment being used is of the type casually thrown together by the service technician, cabinetless, or hopelessly outdated, the shop owner would do well to bury this equipment on the service bench in some remote part of the store. This type of equipment lends no prestige and often might shake the customer's confidence in the service technician to whom he had planned to entrust his radio or video set.

Don't misunderstand us—the amount of beautifully-matched test equipment on a service bench is no infallible criterion of the work capable of being performed in that particular repair shop, but often the customer is far more impressed with such a display of equipment than conditions warrant. It seems to be human nature to seek the external trappings of a successful business and then assume from the visual evidence that here indeed is an A-1 establishment.

There are many service technicians who can perform miracles with a v.t. v.m. and those good little grey cells and more power to him! The point we are trying to make is that if you have invested large sums in modern test equipment don't relegate it to the bench located in the recesses of the shop but have it visible where it can do its bit in promoting "showmanship."

Yes—we radio and TV technicians can learn much by visiting other shops which use "showmanship" and use it to good advantage. Let's give it a try.

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May, 1950

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RADIO & TELEVISION NEWS

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Thousands of TV kits have been sold to the man who likes to build his own equipment. These handymen are ripe for PROTELGRAM, because they can combine it with a TV chassis, get lifesize TV at a reasonable cost.

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Clients who want built-in installations in walls or cabinets are perfect prospects for PROTEL-GRAM. Huge picture size, plus compactness and flexibility, makes it the answer for this type of user.

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PROTELGRAM sells itself to customers who want bigger pictures, but are reluctant to take a trade-in loss. You can now use their present TV chassis, connect it with PROTELGRAM in a cabinet such as shown at (3) left. They get a 234 square-inch picture, $13\frac{1}{2}$ " x 18".

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May, 1950

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RABIO & TELEVISION NEWS



Sylvania's early pioneering in radio testing devices naturally places this company in a position to step ahead in the field of television testing equipment.

So again Sylvania comes through! This time for service dealers everywhere, with a splendid new line of TV Test Equipment. Here are the first two instruments in this line. A new TV Marker Generator will be announced soon. Mail coupon for prices and latest specification sheets.



Sylvania TV Sweep Signal Generator

This compact instrument is equipped with electronically controlled sweep circuits to eliminate the complexities inherent in mechanical type sweeps.

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Presenting latest information on the Radio Industry.

By RADIO & TELEVISION NEWS' WASHINGTON EDITOR

TV COLOR activity zoomed on famed Constitution Avenue, in the electronic labs of the Commission at Laurel, Maryland, and in the labs at Princeton, New Jersey, Madison Avenue, New York City, and San Francisco, California, as the furiously-paced renewed hearing and test sessions began in the nation's capital in the early days of spring.

The second comparative-test session roared in as the curtain raiser for the three-colored event, with both the audience and performers, throbbing with first-night jitters. With a room about twenty-five by thirty feet as the theater, twenty-one receivers as the equipment, and about 125 persons looking, listening, and adjusting, the scene was quite a hectic one. There were miles of overhead cables, crowded folding chairs and swaying viewing heads to add to the confusion and audio bursts from different programs which caused many a flushed face. Reaction to the tests ran from very good to completely inconclusive. From the data that was collected, it did appear, however, as if the technical group was able to gather information that would be very helpful in evaluating the systems demonstrated.

Receivers in the room and also installed in a balcony, included ten-inch, twelve-inch, sixteen-inch and fifteenby-twenty and eleven-by-fourteen inch projection models, all of which were operated simultaneously from antenna setups installed by the color proponents. Signals came from WNBW, operating on Channel 4, for the RCA sets; from WOIC, telecasting on Channel 9, for the CBS receivers; and from WMAL, using Channel 7, for the CTI models. Program material included wedge test patterns, moving horizontally and vertically to simulate camera movements, rotating paddles of various colors moving at varying speeds, keyboard color and black and white patterns, and fixed and live copy. In the fixed setups were cereal boxes, toweling, canned good containers, while the live portions featured dancers, models, and singers attired in bright costumes.

Laboratory signals were also aired to enable the specialists to study such problems as fringing, interference, and ghost reception.

That the test period was a momentous one became strikingly obvious when the audience was surveyed. For among those in the lab room were David Sarnoff, Dr. C. B. Jolliffe and D. F. Schmit of *RCA*; Dr. Frank Stanton and Dr. Peter C. Goldmark of CBS; Dr. Allen B. Du Mont and Dr. Thomas T. Goldsmith, Jr., of Du Mont Labs; Kenneth Norton of the Bureau of Standards; Raymond C. Cosgrove, RMA prexy, and Jack Poppele, prexy of TBA. Canada was also represented with C. W. Browne and W. A. Caton, controller of radio and chief inspector of radio, respectively, of the Department of Transport, as well as W. B. Smith, engineer in charge of the department's broadcast division and W. Hayes of the Canadian Broadcasting Company.

During informal interviews, it was learned that some of the FCC representatives were quite impressed with the demonstrations. Madame Commissioner Frieda Hennock indicated that she felt that the *RCA* system had displayed significant improvement over the setup shown during the first series of comparisons.

AN OFFICIAL VERSION of what the Commission, as a whole perhaps, thought of the tests appeared in quite a frank talk by FCC Headman Wayne Coy, at a conference at the University of Oklahoma. Commenting on the CBS system, for instance, Coy said: "The system proposed by CBS has been in development longer than any of the other two systems and has progressed farther than any of the other two systems. The system has worked well and produces pleasing color pictures of good quality and good color fidelity. At the present stage of development the system must necessarily use a mechanical rotating filter as the means of producing color. Since the mechanical filter must be more than twice the size of the tube, it is obvious that direct view tubes under this system, as a practical matter, can be no larger than ten to twelve inches . . . the larger sixteen to twenty-inch tubes cannot be used with the color wheel. . . . In addition, the CBS system involves a change of transmission standards in two important respects. First, the number of

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lines is reduced from 525 to 405. Secondly, the number of fields is changed from the 60 of black and white television, to 144 for color. CBS urges that the additional intelligence which results from color more than compensates for the loss in geometric resolution which results. However, when a station is broadcasting color programs, in accordance with CBS standards, an ordinary black and white receiver will not be able to receive the program in black and white, unless either a manual or automatic adapter is incorporated into the black and white sets. But that is not color television."

Then the FCC spokesman went on to describe the RCA system, which he recalled was very unsatisfactory in the early demonstrations, but displayed improvement as the recent tests revealed.

"However," he added, "at the present time, the system relies on three tubes to produce the color image. This is costly and a cumbersome procedure, and involves fundamental problems of registration. If the system works out, present black and white receivers will be able to receive color broadcasts in black and white, with no change in the receiver."

"But," he emphasized, "that is not color television either. . . . Getting black and white reception on present receivers from color broadcasts deals only with an economic aspect of the problem; namely, obsolescence of present-day receivers or lack of obsolescence or the cost of preventing such obsolescence as occurs when television programs, which one wants to see, are broadcast in color. . . . One additional fact about the RCA system is that at the present time a color signal will not pass through the coax presently available. The cable width used at present is about 2.7 megacycles and the RCA system requires about 4 megacycles. An RCA signal passing through the 2.7 megacycle coax emerges as a black and white signal."

In commenting on the third system, the CTI, the Commissioner noted that during two attempts demonstrations before the Commission were not successful and that a third was in prospect.

Reviewing the advantages of the six megacycle systems, unveiled during the hearings and recent comparison tests, Coy said that . . . "Certainly a system that would allow black and white reception on present-day receivers would be most desirable. . . . At the present time we do not have such a fully developed system. On the other hand, we do have a workable color system with good picture quality, but it is handicapped by limitations on tube size and the fact that it cannot be received by present black and white receivers without a change."

The answers to these puzzling problems involve many factors, Coy emphasized. He pointed out that the Commission must decide whether it's wise to choose an undeveloped system because its color transmission can be

received in black and white on existing receivers, or choose a workable system whose color transmissions cannot be received on existing receivers in black and white, or permit all of the systems to have more time for experimentation and development.

Then Coy added: "If more time were to be allowed for experimentation and development of all color systems, could the freeze be lifted and permit black and white television to go ahead? Or should the freeze be kept on until it is possible to write engineering standards for color television ?"

The FCC chairman then posed several more key questions. He asked: "If there were to be time for further experimental and developmental work, is there any action, which, taken now, would protect the public presently buying television receivers in a manner which would reduce the obsolescence factor, in the event the Commission decision favored a color system incompatible with present black and white standards? . . . Given the assumptions I have just stated, could it be expected that the industry would immediately build manual or automatic adapters in all receivers? Would the relative small cost for such an addition to a set be worth it, in terms of giving the widest possible latitude for decision, after further experimentation extending perhaps a year or more? . . . Or must we write multiple standards and let the public decide in the market place which is the better system? And is there any way to assure the future use of all color sets if such a proposal is followed?"

The Commissioner then placed these questions before the audience: "Is it a proposal (the multiple-standard proposal) in which the public will willingly accept the risks and costs involved? Or must we reject all of the systems and say that we will proceed immediately with black and white television in both bands and let color wait until another generation or two, and in another, and now unexplored, portion of the spectrum?"

"These are the very difficult questions which the Commission has to face and resolve before it can decide the color television issue," Coy declared. "And," he added, "I am not prepared to answer these questions now."

Commenting on queries as to the hearing's end, in which everyone is so interested, Coy said: "I make no prediction when the hearings will end, or when a decision will be made. I wish the decision could be made tomorrow, but it won't be."

He said that he hoped, but would not predict, that the Commission would be able to resume the processing of applications before the year is out.

CTI, the wallflower contestant in the color race, came to life a few weeks after the official tests, with its own (Continued on page 112)

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COVER PHOTO—Courtesy of RCA

RCA's new "Super-Power Beam Triode" 5831, mounted for testing, is examined by Dr. L. P. Garner, head of the Advance Development Lab of the RCA Lancaster tube plant. The new tube, which can be operated with maximum rated plate voltage and plate input at frequencies up through the "Standard Broadcast Band" and much higher, is being tested at 1,000,000 watts input.



CUSTOM SOUND INSTALLATION

By ARTHUR SHAPIRO

Consulting Engineer, Chicago

Fig. 1. "Wheels-A-Rolling" stage from south end of grandstand. The pylons containing the speakers are at the extreme right and left of the stage.

Design of sound equipment used in the pageant "Wheels-A-Rolling" at Chicago's Railroad Fair.

N designing the sound installation

for the great pageant, "Wheels-A-Rolling," presented at the Chicago Railroad Fair of 1948-49, many unusual conditions had to be met. The tremendous scope of this show can be realized by the great numbers of people and equipment placed on the stage during the one hour and eight minutes of pageantry that told the complete story of transportation in America from early Indian times to the present day. To present this show required a cast of over one hundred actors, 40 engineers, 25 stagehands, stablehands, orchestra, singers and production staff-a total of 250 people who, together with the 94 horses, 35 locomotives and their trains, 32 old automobiles and 100 pieces of horse-drawn equipment, were all controlled from a central point by means of sound, intercommunication, buzzer, light and railroad signal equipment which had to be designed especially for this show.

The audience of 6200 that witnessed this spectacle four times daily was housed in a grandstand that measured 318 feet from stage right to left and 80 feet deep starting at ground level at the stage and rising to 30 feet at the rear of the stands. The grandstand had a metal roof extending from end to end and covered approximately twothirds of the seats. The stage proper was over two hundred feet wide and 100 feet deep with its wings extending either end to make a total play area 500 feet long.

The pageant was historical in tenor with variety and comedy that made for a complete and enjoyable show. Music, especially arranged by Isaac VanGrove, was played throughout the show by an orchestra housed in a special band shell at the south end of the grandstand. The offstage singers were also in the orchestra shell as was the control keyboard for the *Deagan* Celesta-Chime, an electronically amplified chime which will be noted later.

Two narrators, located in a booth at the top of the grandstand, told the story of transportation and gave descriptions and comments of the events as they were enacted on stage. Actors' speaking parts and group singing were picked up on stage by means of concealed microphones which were plugged in under cover of stage action.

The pageant was scheduled for presentation four times daily for the onehundred days from June 25 to October

Fig. 2. Cross-section of the baffles, showing orientation of the speakers.



2nd with performances at 2, 4, 7, and 9 P.M. From this schedule it can be seen that the equipment would be in almost constant service from early afternoon through ten at night. Before the actual show started, there was recorded music to fill the time while the audience found their places. This music was presented for one hour before the 2 and 7 o'clock performances and one-half hour before the other shows. In case of power failure on the grounds, there was a gasoline powered generator as a stand-by to operate the sound system only, so that the show could be presented during the day and to keep the sound operating to avoid panic in the event of power failure during the night shows.

Heavy-duty and trouble-free equipment was a definite requirement. Such items as weatherproof "Twist-Lock" microphone connectors, instant metering of circuits, duplicate wiring and stand-by units were all provided to obviate the possibility of breakdown of the equipment. Since the stage was located on the lake front, the weatherproofing of the equipment was a definite factor to be considered. The high humidity made it necessary to take special care that no space was imperfectly sealed in the equipment that would allow the formation of condensation. On open equipment, sufficient ventilation was required to prevent condensation even when the humidity hovered

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around 98 per-cent. As an indication of the severe humidity, it was not unusual to pick up a tool that had been perfect the night before and find it covered with rust the next morning.

As can be seen from the accompanying drawing, Fig. 7, eleven microphone receptacles were installed, four in the orchestra, one in the room below the orchestra (used for the Celesta-Chime), four on stage and two in the narrators' booth. These receptacles were of the heavy-duty, three wire "Twist Lock" type mounted in weatherproof condulets with spring loaded lift covers. Shure Model 556 Super-Cardioid microphones, set at 250 ohms, were used in all locations except at stage right where a lapel microphone was used in front of the speakers by the actor portraying Abraham Lincoln. This microphone, a high-impedance crystal, required a transformer at the point where it was plugged in to feed the 250 ohm balanced input line.

All microphone lines in the stage area were run underground in steel conduit to a junction box set in the center of the stage apron and then through a two inch conduit under the railroad tracks at the front of the stage to the front of the grandstand where they joined the lines from the orchestra pit. The eleven microphone lines were then brought up to the operating position near the top of the grandstand to a 24 jack patch panel. This panel terminated all the microphone lines and normalled them through to the ten console inputs. The patch panel was installed to permit the instantaneous substitution of an unused console preamp in case of failure of one in use. Inserting a patch cord in the top, or microphone, row of the panel automatically disconnected both sides of the line and transferred the incoming signal to the plug which could then be inserted into the bottom, or console input, row. The eleventh microphone line terminated in a fixed "H" pad. This line was used for the electronically amplified Celesta-Chime. The chime preamplifier fed the signal to the line at 500 ohms and zero level. The front side of the pad fed the signal at 500 ohms direct to either of the loudspeaker systems described below, and the back side of the pad fed through a patch cord to one of the console inputs at 250 ohms



Fig. 3. Ten position, dual channel custom-built control panel in operation. One of the telephone-type intercom masters is visible at the extreme right.

Fig. 4. Block diagram showing essential components of the complete system.



and -55 db. Normally the patch panel operated with only one cord in use as all other circuits were normalled through.

The custom-built console, Fig. 3, contained ten independent preamplifiers feeding through a fader and switch to either of two outgoing channels. Each channel featured independent VU meters, UTC tone equalizers and amplifiers, master faders and output switches. Each preamp position had red and white indicator lights to show which channel was being fed. In operation, the "Red" channel was used for music and the "White" channel was used for speech. This enabled the operator to control the fading of several microphones with a single control. In all there were 37 tubes in the console, three in each preamp and the balance in the equalizer and output section. To eliminate any possibility of hum pickup, all filaments and also all indicator lights were operated on 12 volts d.c. The power supply, mounted within the special weatherproof cabinet, consisted of a heavy duty selenium rectifier for the low voltage and a 5U4G for the plate voltage rectifier. The output of the console was fed through the patch panel

to the line amplifier-driver located in the main amplifier racks in the director's booth. This driver amplifier had provisions for additional high-impedance microphone inputs as well as the 500 ohm low level input. One of these high-impedance inputs was utilized for the director's break-in microphone for use in case of serious accident or emergency on stage. This microphone could break into the system regardless of the status of any other microphones.

The output of the driver amplifier fed four 80 watt booster amplifiers which were limited to approximately 50 watts output each. These boosters were mounted in one rack and were plug connected so that, in case of a failure, a spare could be quickly interchanged. The need for instant changeover was not deemed necessary as each amplifier fed only one-half of the right or left side of the stage, and failure meant only a slight reduction of volume and would not interrupt the show.

The loudspeakers, by far the most important pieces of equipment in the installation, consisted of eight Jensen P15-L cone type speakers specially weatherproofed and treated. These speakers were mounted in special baffles, Fig. 6, and connected two to an amplifier. There are several unique features in the design of these baffles and a study of Fig. 2 will show how each individual motor was oriented to a specific coverage area relative to the grandstand. The pylons in which these units were mounted were not equally spaced and were different distances from the grandstand. This required that the speaker coverage areas be accurately plotted.

Due to the comparatively small difference in angle of each pair of speakers, it was decided that two speakers would be mounted in a common horn and rear enclosure. The loudspeakers had a downward tilt of approximately 10 degrees and the "floor" of the baffles sloped 15 degrees. This angle placed the aural center of the unit at the middle row of the grandstand. These horns were made of three-quarter inch marine grade plywood and finished with flat black paint on the inside so that the speakers were not visible from the audience. The tops of the pylons which supported these large baffles were built with an opening sixteen feet long and four feet high faced with one-quarter inch square mesh screening and treated architecturally with horizontal two by six planks as shown in Fig. 1.

It was evident that with the great variation of distance that the sound had to travel to reach various parts of the grandstand, some control was necessary to drop the level of those speakers operating with the shortest "throw". This was accomplished by inserting a heavy



Fig. 5. Two of the special dual loudspeaker housings before being hoisted into place. Flare of horn and mounting of speaker enclosure can be noted on rear horn.

25 watt "L" pad in the line of each of the two inside speakers of each pylon. In the case of the north pylon, where there was a ladder and platform, these controls were mounted on the speaker enclosures. In the south pylon where the speakers were inaccessable, these controls had to be mounted in a weatherproof box remote from the speakers. Once these controls were set by experimentation and measurement after the installation was completed, they were not changed for the rest of the season. Referring to the block diagram, Fig. 4, it can be seen how these controls were installed. The entire grandstand was covered with a variation of not more than 6 db. This was accomplished by sound pressure measurement and setting each booster amplifier separately and then setting the "L" pads on the individual speakers.



Fig. 6. Dual speaker enclosure showing speakers mounted on individual sections. Note lining of sound absorbent material.

This may seem to be adjusting rather (Continued on page 31)





MICROWAVE TRANSMITTERS



Typical microwave transmitter using a lighthouse oscillator.

First of two articles discussing the generation, modulation, and frequency control of microwaves.

THE compactness and simplicity of microwave transmitters is of great interest to all engineers, particularly so to most "low frequency" engineers accustomed to regarding point-to-point transmitters in terms of large size. The main reason for the reduction in size-for equivalent service -is the large power gain obtained by microwave antennas. It is generally true that the higher the frequency, the more expensive and inefficient the generation of power but, due to the increase in antenna gain, less transmitter power is required. It is therefore possible to operate at higher frequencies -all other factors remaining equalwithout too much loss in over-all system efficiency, provided directional point-topoint transmission is desired.

Since it is more expensive to operate at microwave frequencies (for a given power output), virtually all transmitters operating in this band employ high level modulation, i.e. one r.f. stage is used with the modulating voltage amplified to a sufficient degree to modulate the final at the desired power output level. Thus in considering the design of a microwave transmitter, we are primarily interested in the final stage—all other stages employing conventional techniques.

Microwaves can be generated in a number of ways. The most important of these are the lighthouse triode oscillators, klystrons, magnetrons, and traveling wave tubes. With the exception of the lighthouse triodes, the tubes used at these frequencies are as different in concept and design from conventional tubes as amplitude modulation differs from frequency modulation. The important characteristics of these four methods of microwave generators are:

1. Lighthouse Triodes. Operating range: up to about 3300 mc.; peak power of the order of kilowatts depending upon frequency and duty cycle; average power of the order of watts. This tube requires an external cavity designed by an engineer. This cavity

By J. RACKER

Federal Telecommunication Labs.

is permanent and only the tube must be replaced when it burns out (in contrast to other tubes as will be noted), therefore from a long term viewpoint it is the most economical method. Requires relatively elaborate automatic frequency control systems when a high degree of stability is required, can be pulsed, operated c.w. or amplitude modulated. Difficult to frequency modulate.

2. Klystrons. Frequency range: 900 to 21,000 mc.; power (peak and average) of the order of 5 watts depending upon frequency; efficiency is poor—less than 10%. Cavity is built right into tube and therefore no special cavity design is necessary unless other design problems are involved. Entire assembly must be replaced if tube goes bad. Can readily be frequency modulated, electronically tuned and frequency controlled (a.f.c.). Requires multiple, well regulated power supplies which are relatively expensive and bulky. Cannot readily be amplitude modulated.

3. Magnetrons. Frequency range: 900 to 30,000 mc.; peak power of the order of megawatts depending upon frequency and duty cycle; average power of the order of kilowatts depending on frequency; efficiency is very good—better than 50% depending on frequency. Cavity built right in and entire assembly must be replaced if tube goes bad. Good frequency stability—depends only on expansion of copper. Not easy to



Fig. 1. (A) and (B) Two types of double coaxial resonator circuits. (C) Re-entrant cavity circuit. (D) Quarter-wave choke used as r.f. shorting plug.

a.f.c., amplitude or frequency modulate. Usually used for c.w. or pulsed applications.

4. Traveling Wave Tubes. Frequency range: 500 to 10,000 mc. (and higher); power, c.w., is of the order of hundred watts depending upon frequency; efficiency may be as high as 60%. These tubes are particularly noted for their broad bandwidth. At the present time they are largely experimental and show great promise for the future.

This brief summary of the characteristics of the four major methods of generating microwaves indicates the primary considerations confronting the engineer in designing the transmitter. When the power output, carrier frequency, and method of modulation are specified, the choice of the optimum method of generation is usually apparent though other system characteristics such as bandwidth, frequency stability, and equipment size should also be evaluated before the final selection is made.

Transit Time Effect

It is known that a finite period of time is required for an electron to travel from the cathode to the plate of a tube. This period of time is called the transit time and at low frequency ranges it is negligible compared to time required to evolve one cycle. However, starting at about 500 mc., the time required for one cycle of operation is less than 0.002 microseconds which is comparable to the transit time in most conventional tubes.

This fact invalidates many of the characteristics and design principles governing these tubes since the assumption had been made that the electron stream flowing from cathode to grid traveled under essentially static conditions, i.e., transit time small compared to time variation of tube potentials. However, when the transit time becomes comparable to the period time, then the tube conductance is increased and tube efficiency and output decreased.

This can be best understood by considering the action in a triode with an alternating current applied to its grid. When the grid is at zero (a.c.) potential, the electrons travel from cathode to plate at the quiescent velocity. With the grid at a positive a.c. potential, the electrons are accelerated and pick up energy from the grid in the cathode-to-grid path and are retarded and lose energy in the grid-toplate path. The reverse occurs when the grid is at a negative a.c. potential so that there is no net energy transfer from the a.c. grid potential source to the electrons.

At frequencies where the period is comparable to the transit time, it is possible for the a.c. grid potential to



Fig. 2. Typical loops, probes, feedback circuits, and combinations.

reverse its phase before the electrons reach the plate so that, in the positive grid case, the electrons are accelerated and gain energy in both the cathodeto-grid and grid-to-plate regions. Under these conditions, there is a net energy transfer from a.c. grid potential to electrons, resulting in an increase in input conductance. In a similar manner it can be shown that energy can be transferred from the source of a.c. plate potential to the electrons during a portion of the cycle resulting in a decrease in power output and efficiency.

Another factor influencing the operation of tubes at microwave frequencies is the lead inductance of the tube elements (plate and cathode) and the tube interelectrode capacitance. In many cases the resonant frequency of these L-C networks falls below or within the range of operation. Hence, microwave tubes must be designed with a minimum of lead inductance and interelectrode capacitance.

The requirements of reducing transit-



Fig. 3. Alternate methods of output coupling—probe or loop circuit.

time and interelectrode capacitance are conflicting, necessitating design compromises. Interelectrode capacitance may be reduced by decreasing the physical size of electrodes and increasing the spacing.

However, increasing the spacing increases the transit time, while reducing element size decreases maximum safe power dissipation. Transit time may be reduced by increasing the plate potential, but increased plate potential means greater power dissipation.

It is therefore readily seen that special tubes must be used for microwave frequencies which overcome the limita-



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Breakdown of a typical lighthouse tube oscillator.

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Microwave transmitter-rear view-showing lighthouse oscillator.



Fig. 4. Theoretical and typical power output versus frequency for α simple triode oscillator.

tion of transit time and capacitance noted above or employ entirely different techniques than conventional tubes. Lighthouse tubes fall in the former category, while klystrons, magnetrons, and traveling wave tubes are examples of the latter category.

Lighthouse tubes are designed for

operation in the frequency range of 200 to 3370 megacycles in which coaxial line resonators and elements are used. A typical lighthouse tube is shown on page 7 and as indicated in this picture it has the over-all appearance of a lighthouse—hence its name. These tubes are constructed in this way to permit the use of coaxial elements which fit right over the cathode, grid and plate leads.

This tube has effected reduction in lead inductances by use of disk seals, and employs a parallel plane structure which permits the use of relatively small electrode areas; thereby making possible low interelectrode capacitances while reducing the spacing between electrodes to the point where the transit time is small in comparison with the period at the frequencies indicated.

Design of Lighthouse Oscillators

There are two types of resonator circuits used in conjunction with the lighthouse tube, one is known as the double coaxial resonator, shown in Figs. 1-A

Fig. 5. Typical automatic frequency control circuit for a microwave oscillator.



and B, and the other a re-entrant cavity, shown in Fig. 1-C. As indicated in Figs. 1-A and B, there are two variations of the double coaxial resonator. In one case, the grid to cathode line is completely independent from the grid to plate line. In the second case, in order to conserve space, the grid to cathode line is folded back over the grid to plate line. This does not alter the fundamental principles of the two resonant circuits inasmuch as the inner surface of the grid cylinder acts as the grid side of the grid-anode line while the outer surface forms the grid side of the grid-cathode line. There is no coupling between these two lines through the common conductor since, as indicated in a previous article¹, the depth of penetration of the electromagnetic field at these frequencies is much less than the thickness of the metal.

The shorting plugs, shown in Figs. 1-A, B, and C, are actually r.f. shorts and not d.c. shorts. This is done by using a quarter-wave "choke" shown in Fig. 1-D. Since this element is a quarter wave long and open circuited, it presents a short circuit to the r.f. energy. The spacing between the choke and the wall of the tube is determined by the amount of d.c. voltage applied to the plate of the tube. The smaller the spacing the more effective the "choking" or "shorting" action, since Z_o will be small.

In the double coax circuits shown in Figs. 1-A and B, both grid-cathode and plate-grid lines have shorting plugs which can be varied to tune over the desired frequency range. The re-entrant cavity circuit, shown in Fig. 1-C, is similar to the "folded back" gridcathode line. However, this circuit is not tuned in the same manner that is used in the double coax resonator. Instead of terminating the tuned circuits in "shorting plugs," the grid line is made one half wavelength long and mounted directly on the flange of the tube. The length of the grid line is fixed for any given setup, and thus becomes the frequency determining factor in the circuit.

The shorting plug in the re-entrant cavity, which is mounted on the plate line, does not affect the resonant frequency of the circuit, but functions primarily to optimize conditions for oscillation with a given grid cylinder rather than to change the frequency.

The double coax and re-entrant cavity circuits are basically degenerative and the amount of feedback provided by the interelectrode capacity is insufficient for oscillation, if care is used in shielding the resonant lines from one another. Probably the most frequent cause of insufficient shielding is poor grid contact design. These lines should be slotted to permit a spring contact. Similarly

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the plate line should be machined and slotted so that it grips the plate cap of the lighthouse tube firmly.

There are a number of feedback loop or probe arrangements that can be used to sustain oscillations. Some of these are shown in Fig. 2. A loop is usually placed at a voltage minimum and a probe at a voltage maximum. When broad tuning is desired, a combination of probe and loop feedback elements may be used to cover the entire frequency range.

It will be found that the frequency and efficiency of the oscillator will depend to some degree upon the position (determining phase of feedback) and size (determining the amplitude of feedback) of the probe or loop. A trial and error procedure is used to determine the optimum feedback circuit that provides maximum output and efficiency.

The output of the oscillator is picked up by either a probe or loop as shown in Fig. 3. The exact position of the probe or loop will depend upon the impedance of the load. Again the best procedure is to vary the position until maximum output is obtained across the load. This effect is similar to tapping the coil of a tank circuit to match a given impedance.

The diameters of the lines are determined by the requirements of the oscillator. Using standard theory, the desired Q of the tank circuit is determined considering both the effects of the tube and the load. The Q of a resonant length of coaxial line was given³ in a previous article as:

$$Q=rac{2\pi f_o\sqrt{LC}}{2lpha_T}$$
 (1)

This equation can be simplified by assuming that the attenuation due to the dielectric, α_a , is negligible, in which case $a_T = R_1/2Z_0$, and Eqt. (1) becomes:

$$Q = \frac{2\pi f_o Z_o}{R_f V_T} \qquad . \qquad . \qquad . \qquad . \qquad (2)$$

since $V_T = 1/\sqrt{LC} = 3 \times 10^{10}$ cm./sec. for air. Further simplification of this equation can be obtained³ by writing Z_0 in terms of D/d and the r.f. resistance² in terms of D, d, and f. Doing this we obtain:

$$Q = 8.39\sqrt{f} DHR_1 \times 10^{-2} \qquad (3)$$

where D is in cm., f in mc., R_1 is the ratio of the d.c. resistance of the conductor used to the d.c. resistance of copper, and H is defined as:

$$H = 3.6 \left(\frac{D+d}{d}\right) \log_s \frac{D}{d} \quad . \quad . \quad (4)$$

Eqt. (4) is plotted in Fig. 7. With the help of this figure and Eqt. (3), the Q of a coaxial line for a given D/dratio can readily be determined.

For the purpose of mechanical simplicity, the plate line diameter is usually made so that it will just slip over the plate cap. The diameters of the grid and cathode rods are then calculated based on this parameter and the Q required. Of course it is desirable to use rods that are of the same diameters as the caps provided or slightly larger.

Earlier in this article it was noted that there is a definite relationship between interelectrode capacitance, electron transit time, and maximum plate dissipation in a tube design, i.e., reducing interelectrode capacitance and transit time also causes a reduction in maximum plate dissipation. Due to the various mechanical factors involved, there is a maximum theoretical (at least at the present time) continuous power output than can be obtained from a single triode employed in the conventional manner. This maximum value has been plotted in Fig. 4 as a function of frequency.

The continuous power output obtained from commercially available tubes is less than this theoretical value. Fig. 4 plots a typical power output versus frequency curve of a 2C43 lighthouse tube operating in a re-entrant cavity. The outputs obtained from other tubes operating with different potentials can be obtained from the tube manufacturers.

At microwaves, where considerable spectrum is available, it is frequently convenient to use greater r.f. bandwidth for greater efficiency and power. This can be done by using one of the



Fig. 6. Plate-pulsed oscillator performance, power output vs. frequency. Pulse rate, 1000 per second; duration, 1 microsecond,





several pulse modulation techniques available such as pulse time modulation (PTM), pulse amplitude modulation (PAM), or pulse frequency modulation (PFM)⁴. When the transmitter is pulsed, a much higher peak power can be emitted for a given maximum average power.

It is known that the average value of a pulse series is equal to⁵:

(Continued on page 30)

Fig. 8. Another type of microwave automatic frequency control circuit.



ENGINEERING DEPT.

Dr. Goubau at the Signal Corps Engineering Laboratories with a model of the simplified "G-String" transmission line which he invented.

SURFACE WAVE TRANSMISSION LINE

By GEORG GOUBAU Signal Corps Engineering Labs.

PAPER published by Sommerfeld in the year 1899 describes a surface wave guided by a cylindrical conductor of finite conductivity. This wave differs from the wave mode usually excited on long wire antennas in two respects: first, it suffers no radiation loss along its path and second, the field distribution depends to a high degree upon the conductivity of the wire material; and the wave would not exist at all if the conductivity were infinite. Though there is a solution of Maxwell's equations in that case, the solution lacks physical reality since the power transmitted by the wave is infinite if the current in the wire is finite.

Sommerfeld's paper has seen little consideration in the modern literature and there are only a few books in which it is quoted. One of these books is Stratton's well-known "Electromagnetic Theory". It may have been doubtful to many a physicist whether this surface wave can be excited, because not every special solution of Maxwell's equations is realizable. Sommerfeld's wave is a plane wave and therefore the wave could be part of an asymptotic solution, which is valid at a large distance from the power source. In that case, only an infinitesimally small amount of the total power would be converted into the surface wave. To our knowledge nobody has proven as yet that Sommerfeld's surface wave can be excited with finite amplitude. Though we are convinced that this can be done, there are two facts which will make it difficult to obtain good efficiency. The field extends

A single insulated wire with special funnel-shaped terminals provides very efficient h.f. transmission.

very far from the wire and has a rather complicated shape. An effective launching device would have to be very large, otherwise the radiating wave mode would become predominant.

ELECTRIC FIEL

The reason we were interested in Sommerfeld's wave for a long time is the theoretically low attenuation of this wave. Transmission lines using this wave mode should have less insertion loss than coaxial cables and—provided the diameter of the wire is sufficiently large—even smaller loss than rigid wave guides. This fact made us think about the possibility of concentrating the field closer to the wire; thus, the excitation of the wave would become simpler and besides, the clearance around the wire required for an un-





disturbed propagation could be kept smaller. A study of cylindrical surface waves showed that the extension of the field of a surface wave depends upon the phase velocity. The more the wave is urged to travel with lower speed, the more the diameter of the field shrinks. A reduction in phase velocity of less than one per-cent is in general sufficient for obtaining reasonable dimensions of the field. Such a reduction can be achieved by a dielectric coat on the wire or by other modifications of the conductor surface like the application of a thread on the wire. The wave mode guided by a conductor, the surface of which is modified as mentioned, differs from Sommerfeld's wave not only in the extension but also in the structure of the field. Both wave modes are mathematically described by Hankel functions; however, in the case of the plain wire, the argument of these functions is complex, while it is purely imaginary in the other case, at least if the losses are disregarded. And here we have an important difference: the finite conductivity is no longer an essential provision for the existence of the non-radiating wave mode.

Last summer we started experiments with this wave mode on dielectric coated and threaded wires. From the very beginning we observed that this wave mode can be easily excited with an efficiency up to 90%. Furthermore, the experiments showed that the wave mode is very stable with regard to a sag of the wire or to small bends. The meas-

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ured loss of transmission lines based on this wave mode was in very good agreement with the theoretical expectations.

Before discussing the setup and the results we would like to say a little bit more about the wave mode. The wave mode is a radially symmetrical transverse magnetic mode with a small longitudinal electric component. Fig. 2 shows the electric field lines for a case in which the wavelength is 100 times the wire radius and the phase velocity is reduced by 10%. In general a much smaller reduction of phase velocity is used. We assumed here a large reduction in order to get more field lines on the drawing. All electric field lines start on the wire and return to the wire. The magnetic field lines are circles around the wire.

Fig. 1 shows how the field decreases with the distance from the wire. The ratio of the magnetic field strength at a distance from the wire to the magnetic field strength at the surface of the wire is plotted versus the distance, measured in multiples of the wire radius. Both scales are logarithmic. The broken line indicates a 1/r decrease as it would be present in the case of a plain wire with infinite conductivity. In this case the phase velocity would be equal to the velocity of light, and as mentioned before the power would be infinite if the field strength were finite. The unbroken curves show how the field decreases if the phase velocity is reduced by 1, 2, 5 and 10%. The curves follow first the l/r decrease and approach at larger distances an exponential decrease. The smaller the phase velocity is, the earlier the exponential decrease begins.

Fig. 3 is a schematic sketch of a surface wave transmission line. The launching device used in the experimental work consisted of a metal horn connected to the outside conductor of the coaxial feed lines. The inner conductor of that line is connected to the surface wave guide consisting in most of the experiments of a normal enameled wire. The transmitting and receiving ends of the line are alike. The dimensions given in Fig. 3 refer to the curves of Figs. 5 and 6.

The loss of such a surface wave transmission line consists of three parts: the conductivity loss in the wire, the loss in the dielectric coat and the loss, effected by the launching device, due to a partial excitation of the radiating mode. The first two parts, conductivity loss and dielectric loss, can be calculated from the field distribution of the wave. They are proportional to the length. The launching loss is independent of the length of the line and can be determined with fair accuracy by the following consideration. At the receiving end that portion of the wave energy will be received which travels within the area of the aperture of the horn. The wave energy outside this area will be lost. The ratio between received energy and the total energy determines the efficiency of the receiving horn. Because of the reciprocity theorem the efficiency of the transmitting horn must be the same as the efficiency of the receiving horn.

The loss of the transmission line shown in Fig. 3 was calculated for various frequencies and Fig. 5 is a plot of the results. The length of the line is 120 ft., the diameter of the wire 0.2 cm., and the thickness of the enamel layer 5.0 x 10⁻³ cm. The dielectric constant is assumed to be 3 and the power factor $8 \ge 10^{-3}$. The opening of the horns is 13". The conductivity loss L_c and the dielectric loss L_i increase with frequency while the launching loss L_h decreases. The total loss L_{tot} which is the sum of the three components L_c , L_i and L_h has a flat minimum of about 2 db. The loss was measured for three frequencies. For 1600 mc. it measured 2.2 db.; for 3300 mc., 2.3 db.; and for 4700 mc., 4.5 db. While the deviation between theory and measurement for the lower frequency is comparatively small, it is pretty large for the high frequency. The main reason for this deviation lies in the fact that the angle of the horns was too large for the high frequency and thus the efficiency of the horns was much smaller than expected. For comparison, the approximate loss of 120 ft. of RG-8/U cable is 13 db. for 1600 mc., 22 db. for 3300 mc. and 30 db. for 4700 mc. Though these data refer to a very good cable the losses are about 6 to 10 times as high as the losses of the surface wave transmission line.

The effect of the thickness of the dielectric coat on the efficiency of such a transmission line is demonstrated by curves shown in Fig. 6. Length of the line, diameter of the wire and dimensions of the horns are the same. Only the thickness of the dielectric coat is assumed to vary. The curves refer to a frequency of 3000 mc. For a very thin







Fig. 2. Electric field lines where $\lambda = 100 \ \alpha$ and v/c = 0.9



Fig. 3. Schematic sketch of a surface wave transmission line.



Fig. 4. Calculated attenuation for 2 cm. wire with $\frac{1}{2}$ mm. polystyrene coating.

dielectric coat the insertion loss is large because of the low efficiency of the horns. There is an optimum thickness for which the loss is a minimum. For larger thickness of the dielectric coat the loss rises slowly because of the increasing conductivity and dielectric losses. A curve like this has been verified. The same horns and the same wire length (120'), but a wire of 0.26 cm. thickness were used (this was the only available bare copper wire). The measurements were made at a frequency of 2600 mc. The wire was outdoors for (Continued on page 29)

Fig. 6. Losses for the line of Fig. 3 for varying thicknesses of dielectric coating.



A NEW R-) OSCILLATOR CIRCUIT

By MILTON H. CROTHERS Electrical Engineering Dept., U. of Illinois*

Oscillator using a 12AU7 and operating near 9 kc. Note compactness of unit.

Development of an R-C oscillator having greater stability and fewer parts than conventional units.

TACUUM tube oscillators are truly fundamental to the art of modern communications. Vacuum tube oscillators may be grouped in two general classes depending upon the class of operation of the amplifier stage, that is, linear or nonlinear. The first vacuum tube oscillators depended upon nonlinear operation of the vacuum tube to locate and determine the steady state amplitude of oscillation. Nonlinear operation of the amplifier tube must introduce distortion components of the fundamental signal frequency. These distortion components are attenuated by a highly selective tuned circuit.

Resistance-capacitance oscillators have become very popular within the last decade. This paper will introduce a novel form of this oscillator, making use of the standard R-C network and a new amplifier, and having several marked advantages over circuits which have been used to date.

Need for Nonlinearity in an Oscillator

Let the general oscillator be described as an amplifier section and a feedback section. The gain of the amplifier will be represented by the symbol K and the transmission of the feedback section will be represented by the symbol β . Either one or both of these sections will include some form of a selective circuit so the gain through the sections will be a maximum at only one frequency, as indicated in Fig. 1.

• From the thesis submitted in partial fulfillment of the requirements for the degree of Master of Science in Electrical Engineering to the Graduate College of the University of Illinois 1949. 'A Study of Nonlinear Circuit Elements'.

When the input signal (E_1) and the output signal (E_3) are identical the loop may be closed to form an oscillator. If the net gain around this loop is greater than unity the oscillations will build up in amplitude with each succeeding cycle. The gain around the loop must be self adjusting in order that the oscillator maintains some particular value of oscillation amplitude. The steady state amplitude in an oscillator is defined as that signal level which just makes the loop gain unity. An amplifier operating in a nonlinear region varies its effective gain somewhat inversely with signal amplitude and thereby controls the oscillation amplitude.

Nonlinearity and Circuit Selectivity

An oscillator using any nonlinear element which is instantaneous in nature generates distortion components. The



Fig. 2. (A) Tuned L-C circuit. (B) Tuned R-C network.



relative distortion depends upon the degree of nonlinearity over the operating range. Attenuation of these distortion components depends upon the selectivity of the circuit. Highly selective circuits permit highly nonlinear operating regions in the amplifier.

The resistance-capacitance network shown in Fig. 2 may be used as a selective network. The Q of the R-C network is only $\frac{1}{3}$ when equal values of resistance and capacitance are used. A highly selective tuned circuit is compared with the R-C network in Figs. 3 and 4.

Very little attenuation is offered to distortion components by the poorer selectivity of the R-C network. For this reason there must be control of the loop gain by a method which is not an instantaneous nonlinear element.

Nonlinear Elements Which Are Not Instantaneous

Most of the resistance-capacitance oscillator circuits use thermally responsive nonlinear devices. Generally these elements are low current 115 volt tungsten lamps. Low current fuses are also suitable for higher frequency signals. These devices have a positive temperature-resistance coefficient. Typical characteristics of a number of common nonlinear elements are given in Figs. 5 and 6.

The lamps show linear operation within each cycle for frequencies as low as ten cycles per second. They will not introduce distortion components when operated over the audio range. The small fuses show some instantaneous nonlinearity at frequencies below 100 cycles and therefore are not suitable for use at frequencies lower than 1000 cycles.

A fundamental and useful oscillator circuit using class ${}^{\prime}A_{1}{}^{\prime}$ amplifiers with an *R-C* feedback network is shown in

Fig. 7A. Pentodes are normally used but triodes are shown to illustrate the basic circuit.

A great amount of inverse feedback is included to stabilize the circuit. The inverse feedback circuit includes a nonlinear element R_x which controls the gain inversely as the signal amplitude varies. The gain of this amplifier is indicated in Fig. 13.

The condition required for steady state oscillation is that the loop gain is unity, that is, E_1 and E_3 are identical. This will lead to two general expressions:

$$K = 1 + \frac{R_1}{R_2} + \frac{C_2}{C_1}$$
. (1)

Freq. =
$$\frac{1}{2 \pi \sqrt{R_1 C_1 R_2 C_2}}$$
. . . . (2)

Adjustment and selection of circuit elements and tube coefficients should be made so the action of R_x limits the oscillation amplitude to class A, operation. The nature of control obtained from R_x depends upon the heating by the a.c. signal current. It is to be noted that d.c. quiescent current must also pass through R_x and thus hinders the control action. Oscillator circuits of this general type have been used with great success in a large number of applications over frequency ranges of 1 to 200,000 cycles. The distortion in most cases is of the order of 2 per-cent or less.

These are the general characteristics desired of the amplifier section selected for this oscillator circuit:

1-Gain is stable and constant as a function of supply voltages, component values (in general) and tube changes.

2—Wideband frequency response and minimum phase shift over the working frequency range.

3—Low output impedance compared to values of elements used in the feedback network.

4—Nonlinear element controlling the gain inversely with signal amplitude. This element must not introduce distortion and the signal level must be within the class A_1 region.

The New Amplifier Circuit

The circuit devised by the author and shown in Fig. 7B complies in general with the specifications listed above and offers a number of advantages. A fewer number of circuit components are required with this new " π coupled" amplifier circuit. There is also one fundamental advantage in the manner in which the nonlinear element operates and responds to the signal level. These items will be discussed in detail.

The " π coupled" amplifier is that part of the circuit within the dotted lines and uses one dual triode tube, three resistors and one nonlinear element. This is a total of four circuit elements



Fig. 3. Frequency response of networks for a Q of 1/3 and a Q of 10.

as compared to the nine components required by the amplifier in Fig. 7A. This smaller number of elements offers decided advantages where cost, space, and weight are to be considered. This is the major reason for advancing this new circuit.

The second advantage results because the nonlinear element R_x does not carry quiescent current in this new circuit. The element can respond more readily to the a.c. signal component in the circuit and offers a greater sensitivity to signal amplitude. This greater sensitivity offers some compensation for the less stable amplifier characteristics. It is true that any of these circuits could be arranged to impress only the a.c. signal across the element R_x but this new circuit provides such isolation within the basic form. A total of three variations will be given for this circuit. These variations differ only in the manner of balancing out d.c. from the nonlinear element.

The amplifier shown in Fig. 7A has low output impedance because the inverse voltage feedback reduces the output impedance below the value of the amplifier without this inverse feedback. The output impedance of the new amplifier is higher because there is no inverse feedback applied. A value of 5-10 k. ohms is typical of this output imped-







Fig. 4. Selectivity of a network for a Q of $\frac{1}{3}$ and a Q of 10.

ance. The value of R_1 in the feedback section might be corrected to allow for this impedance.

Examination of the circuit for the " π coupled" amplifier reveals that the first stage is operating as a grounded plate amplifier and the gain is therefore less than unity:

The gain of this stage is reasonably constant with respect to changes in supply voltages.

Fig. 6. Characteristics of lamps.



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Fig. 7. (A) Fundamental circuit using class A_1 amplifiers with an R-C feedback network. (B) New " π coupled" circuit has advantages over (A).



Fig. 8. Variation in amplifier gain with respect to signal level.



Fig. 9. Variation in amplifier gain with respect to plate voltage.



Fig. 10. Frequency stability of typical oscillator plotted against variations in plate voltage.





This first stage is coupled to the second stage by means of the π type network between the cathodes. For this reason the name of " π coupled" amplifier has been coined for the circuit. The second stage operates as a grounded grid amplifier, the gain of which is given as:

Interaction between these stages through the π coupling network gives a net gain different than that of the product of K_1 and K_2 above. The net over-all gain has been calculated and is given in this form:

where A, B and C are:

$$A = \mu_1 \ (\mu_2 + 1) \ R_{L_2} \ R_{k_1} \ R_{k_2} \ . \ . \ (6)$$

$$B = (R_{L} + r_{p}) [r_{p}R_{k_{1}} + r_{p}R_{k_{2}} + (\mu + 1)R_{k_{1}}R_{k_{2}}] + r_{p}R_{k_{1}}R_{k_{2}}(\mu_{2} + 1) .$$
(7)

Loading of the amplifier by the feedback network has not been considered in

Fig. 12. Distortion in 400 cycle 6SN7 R-C oscillator with (A) $R_x \alpha 4 w$. lamp and $E_{bb} =$ 200 v., (B) $R_x \alpha 5$ ma, fuse and $E_{bb} =$ 250 v.





the calculation of amplifier gain. It is assumed that this may be neglected in most cases by proper selection of the elements in the feedback network.

Typical Operation

Most of these oscillator circuits are used with equal values for the resistances in the feedback network. The gain which is required is then found to be three. The dual section triodes 6SN7 and 12AU7 may be used with these approximate circuit values:

$$R_{k1} \equiv R_{k2} \equiv 2,000 \text{ ohms.}$$

 $R_{L2} = 10,000$ to 15,000 ohms selected for oscillation level.

 $R_x =$ one 115 volt, 6 or 7 watt tungsten lamp.

The feedback network values are selected for the frequency desired with these general limits suggested:

$$R_1 = R_2 =$$
 values ranging from 50k ohms to 10 megohms.

$$C_1 = C_2 = \min$$
 values of 100 $\mu\mu$ fd.

These feedback network values will permit operation over and slightly beyond the audio frequency range.

A number of oscillators have been constructed using the basic circuit with fixed elements in the feedback network. These oscillators have been built within the general circuit of some projects to provide various signals at one frequency. Another oscillator has been constructed using a pair of switches to select any one of a number of fixed frequencies. Continuously variable oscillators have also been constructed using either variable resistors or variable condensers in the feedback circuit.

From a number of these different oscillators the curves given in Figs. 10, 11 and 12 have been recorded to illustrate the general operation of this circuit.

Long time measurements are not indicated for frequency stability for this oscillator does not seem to fit in this general class. The variation in frequency with respect to supply voltage is favorable on a short time test. The frequency will be sufficiently constant for most applications in mind for the circuit.

Measurements on a number of different oscillators indicated that an output voltage with 1 per-cent or less total distortion can be obtained operating near the oscillation threshold. Additional measurements indicate that lower distortion was obtained, for any given output voltage level, when the plate supply voltage is increased and the amplifier gain readjusted for the same output level. In making the observation the gain was adjusted by varying the value of $R_{\pm 2}$. In other circuits the (Continued on page 24)



By ROBERT C. MOSES Sylvania Electric Products Inc.

Various types of compensation may be used to give an amplifier with essentially flat amplitude and phase characteristics over the entire video band.

IGH definition television systems are required to convey information transmitted at rates up to 8,700,000 picture elements per second. The ultimate resolution obtainable in the reproduced image is determined by the ability of the equipment as a whole to respond to the maximum rate of transmission. The latter, in turn, is directly proportional to the bandwidth of the circuits through which the video signal must pass. With the present 525 line, 30 frame transmission standards, it can be shown that adequate response up to a maximum frequency of 4.35 megacycles is necessary for faithful reproduction of the finest detail in the picture. On the other hand, satisfactory rendition of backgrounds and of very large objects requires that the system be capable of reproducing one picture element having dimensions approaching those of the entire scanned area. Proper response of the system down to d.c. is therefore indicated. Since transmission of d.c. in the system is a subject beyond the scope of this article, the following discussion will be limited to amplifiers responding to frequencies down to about 30 cycles.

The necessity for handling such a wide spread in rates of transmission imposes severe requirements upon the video amplifier channel. Not only must the amplifier be capable of passing the entire frequency range from 30 cycles to over 4.0 megacycles with a minimum of amplitude discrimination, but also

the transmission time through the system must be as nearly constant as possible for all frequency components within these limits. The latter requires that the curve of amplifier phase shift versus frequency be essentially linear up to the maximum frequency present in the video signal.

The usual resistance-coupled audio amplifier falls far short of these requirements, since both the amplitude and phase characteristics deteriorate rapidly above about 10,000 cycles and below 70 cycles. Some form of compensation for both amplitude response and phase shift is obviously necessary. There are several methods by which such compensation may be accomplished, and for purposes of discussion, the low and high frequency considerations will be treated separately.

Even if the amplitude response of the conventional pentode resistance coupled amplifier were satisfactory at low frequencies, the phase characteristics below 60 cycles would be troublesome. Excessive phase shifts at low frequencies may be attributed to three sources:

1. The time constant of the grid coupling network

2. The screen circuit time constant

3. The cathode biasing network These factors are indicated in the equivalent circuit of Fig. 2A. The plate supply terminals and d.c. connections are omitted for clarity.

The reactance of the coupling capacitor C_c at low frequencies introduces a leading phase shift and a drop in amplitude in the component E_{\circ} of output voltage across the grid resistor of the following stage. These effects become serious below the frequency at which the reactance of C_c exceeds about $\frac{1}{8}$ the resistance of R_{g} , and may be sufficiently great to prevent proper reproduction of square waves of 60 cycles base frequency unless large values of C_c and R_g are used. The magnitude of the phase angle is given by:

$$\varphi = -\tan^{-1} \frac{1}{2\pi f C}$$

 $\varphi = -\tan \frac{2\pi f C_c R_g}{2\pi f C_c R_g}$ and the relative amplitude characteristic by:

$$\frac{E_0}{E_i} \equiv \cos \varphi$$

where E_o and E_i are the output and input voltage respectively.

While it is theoretically possible to obtain values of C_c and R_g which will keep the frequency of negligible phase shift below 30 cycles, large values of coupling capacitors introduce additional stray capacitances which may affect the high frequency characteristics of the amplifier. At the same time, the maximum value of the grid resistor R_g is usually limited by gas and grid current considerations in the following tube. It is therefore customary to assign reasonable values to C_{c} and R_{g} , and





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Fig. 2. (A) Low frequency equivalent circuit. (B) Equivalent circuit with low frequency compensation.



Fig. 3. (A) High frequency equivalent circuit. (B) Shunt peaking network.

provide compensation for the resulting phase shift at some point in the system.

The finite time constant of the screen resistor-capacitor combination R_*C_* causes a shift in phase and a drop in amplitude at low frequencies in exactly the same way as the grid coupling network, and the extent of these deficien-

cies may be specified in the same manner. However, the phase angle and departure from flat amplitude response introduced by R,C, can be kept small with practical component values, and their effect need not be compensated if:

$R_s C_s \geq 0.077$

where R, is in megohms and C, is in microfarads, a condition readily achieved in most cases. If the above holds, the phase angle due to R_*C_* will not exceed 4° at 30 cycles.

The impedance of the cathode bias network $C_k R_k$ results in degenerative effects and excessive phase shifts at frequencies below that at which the reactance of C_* exceeds about 1/10 the resistance of Rk. In many cases, prohibitively large values of C_{k} are required if the frequency of near-zero phase shift is to be comparable with the frame repetition frequency. Compensation for low-frequency discrepancies due to $C_k R_k$ can be provided in much the same manner as for the grid coupling network; however, the design constants for cathode-bias impedance compensation are different.

Despite the fact that there are several arrangements for compensating the effects of the grid coupling and cathode biasing networks, one simple form of correction circuit has been widely used. This consists of a parallel resistor and capacitor in the plate circuit of the stage to be compensated, as shown in Fig. 2B.

In this circuit, the additional components $C_d R_d$ in series with the plate load of the tube introduces a compensating phase shift which causes the output voltage E_{\circ} to lag in phase by the same amount as the leading phase angle produced by either $C_c R_g$ or $C_k R_k$. At the same time, the reactance of C_d causes a rise at low frequencies in the effective load impedance presented to the tube. The resulting increase in gain counteracts deficiencies in amplitude response. It is usually not possible to satisfy the conditions for correction of both grid coupling and cathode biasing network effects in one such correction circuit.

Fig. 4. Amplitude characteristic of the two shunt peaking networks (Fig. 3B).



An entirely satisfactory degree of compensation for either can, however, be achieved through a suitable choice of correction parameters.

Compensation for Grid Coupling Network

For a satisfactory degree of compensation for phase shifts produced by the grid coupling network, the following relations must hold:

1. The time constant of the plate load resistor and compensating capacitor must equal the time constant of the grid coupling network.

$C_d R_L = C_c R_g$

2. The resistance of the decoupling resistor must be at least ten times the reactance of the compensating capacitor at the lowest frequency for which correction is required.

$R_{d} \geq 10/\omega C_{d}$

It should be noted that this type of compensation for deficiencies in the grid coupling network will not give perfect correction to very low frequencies or to d.c. except in the theoretical and practically unattainable condition of infinite decoupling resistance R_d . For a fixed value of C_d , the lowest frequency for which compensation is exact is inversely proportional to the value of R_d .

The usual procedure in compensating a stage for the effects of the grid coupling network is to select a suitable value for C_d , say 10 microfarads. This, in combination with the plate load resistor as determined from high frequency considerations, gives the time constant to be obtained in the following grid circuit. The decoupling resistor R_d is then made as large as possible consistent with obtaining the required plate voltage at the tube.

Compensation for Cathode Bias Network

Correction for phase shifts and amplitude discrimination due to the cathode biasing network can be effected in a similar manner to the above, except in this case, exact compensation can be achieved down to d.c. with practical circuit values. Perfect correction will be obtained if:

1. The time constant of the compensating network is made equal to that of the cathode bias network.

$$C_d R_d = C_k R_k$$

2. The decoupling resistor is made equal to the product of the midband stage gain and the value of the biasing resistor.

$R_{d} = \overline{A} R_{k} = G_{m} R_{L} R_{k}$

where G_m is the tube transconductance.

Compensation for All Low-Frequency Effects

It will not usually be possible to obtain values for the correction circuit, such that the requirements for phase

and amplitude compensation of both grid coupling and cathode bias networks are simultaneously satisfied. For correction of all low-frequency effects in a multistage amplifier, the preferred procedure is to make the grid circuit time constants as large as possible in stages having a cathode bias network, and compensate the latter in each plate circuit. Then, for every two or three such stages, a stage having no cathode bias network is provided, and the combined phase shifts of all grid couplings compensated in the plate circuit of this stage. This arrangement operates quite satisfactorily provided the total phase shift to be corrected is smaller than about 20° at 60 cycles. This figure is readily achieved if the time constants of the uncompensated grid coupling networks are from 8 to 15 times greater than the period of the lowest frequency to be passed.

For adequate vertical resolution of a 525 line picture, the maximum delay 1 varitime phase shift \times frequency ation tolerable at the low end of the video band is on the order of 850 microseconds. With two video stages involved, a phase shift of 4 or 5 degrees per stage at 30 cycles can be considered allowable for design purposes. The amplitude response should be flat within a few tenths of a decibel to well below 30 cycles if the above requirement is to be met. Low frequency phase and amplitude characteristics of a typical twostage video amplifier with compensation are shown in Fig. 1.

Tube interelectrode capacitances and stray capacitances associated with circuit components and wiring have a shunting effect which produces excessive phase shifts and departure from linear amplitude characteristics at the higher video frequencies. For a fixed value of plate load resistance, the absolute magnitude of these capacitances becomes the limiting factor on the top frequency f_o , at which satisfactory performance can be obtained. Referring to the high-frequency equivalent circuit of Fig. 3A, it can be shown that a phase shift of 45° together with an amplitude drop of 3 decibels will occur when:

$$f_o \equiv \frac{1}{2\pi C_t R_L}$$

where C_t is the total shunting capacitance, and R_L is the plate load resistance (Fig. 3A). For half-power (3 db.) response to 4.5 megacycles, and a total shunt capacitance of 25 $\mu\mu$ fd., (a reasonable figure for a two-terminal interstage) the above equation indicates that the load resistor must be 1400 ohms. The stage gain is given by:

$$\overline{\mathbf{A}} = G_m R_L$$

For a tube with a transconductance of 9000 ohms the stage gain will be 12.6.



Fig. 5. Time delay characteristics of the two shunt peaking networks (Fig. 3B).

The above considerations lead to a "gain bandwidth" (hereafter abbreviated G.B.) rating for video amplifier tubes, in which the input and output capacitances are compared with the transconductance. For tubes of the same type, coupled by a two terminal network, the input and output capacitances add directly, and the G.B. factor F' is:

$$F' \doteq \frac{G_m}{2\pi (C_1 + C_2)}$$

where C_1 and C_2 are input and output capacitances respectively.

With four terminal and certain filter couplings, the input and output capacitances are isolated from each other. The G.B. factor F'' in this case is:

$$F'' = \frac{G_m}{2\pi \sqrt{C_1 C_2}}$$

Gain-bandwidth factors for several common tube types are given in Table I.

The G.B. factor has the dimensions of frequency, and denotes the bandwidth at which the voltage gain of an ideal stage becomes unity. Thus, a type 6AC7 with a two-terminal interstage coupling would have a gain of 1 at a half-power bandwidth of 90 megacycles; conversely, with a 1 megacycle bandwidth, the stage gain would be 90 times. These figures take into account the tube capacitances only, and do not include the unavoidable strays. (A) C1 C2 C2 C1 C2 C1 C2 C2 C2 C1 C2 C2 C2 C1 C2 C

Fig. 6. (A) Series peaking network. (B) Two section filter coupling.

From the above discussion, it is apparent that a tube having high G_m and low interelectrode capacitances (largest G.B. factor) is the most desirable for video amplification. At the same time, the figures given represent the ultimate performance of which the tube is capable, and will not be realized in practice because of the presence of external circuit capacitances which tend to restrict the bandwidth. Furthermore, the G.B. factor is based on half-power bandwidths, and for television video applications, which may involve several amplifier stages, a much more linear phase and amplitude characteristic in each stage is necessary. The G.B. factor does, however, serve to indicate the rel-





DEPT.







Fig. 9. (A) Modified shunt-series peaking. (B) Cathode peaking.

ative merit of the various tube types in wide-band amplifier applications.

High Frequency Compensation

It is possible to approach more nearly the theoretical performance indicated by the G.B. factor for any tube type through the use of one of a number of two or four terminal interstage coupling networks. Such networks may provide nearly complete phase and amplitude compensation up to a top frequency F_{max} which equals or exceeds

$$\frac{1}{2\pi C_t R_L}$$

This is to be contrasted with the uncompensated case where the departure from ideal phase and amplitude characteristics at the top frequency is very much greater than can be tolerated. In the following discussion of several such coupling networks, the basic circuit parameter for design purposes is the reactance of the total shunting capacitance C_t at the top frequency F_{max} . The figures of merit given in each case apply to the network itself, and indicate its midband voltage gain in relation to the uncompensated case.

Two Terminal Networks

Shunt Peaking, case 1 (Fig. 3B): m = 0.50 $C_i = C_1 + C_2$ $R_{L} = rac{1}{2\pi \ F_{max} \ C_{t}} L_{o} = 0.5 \ R_{L}^{2} \ C_{t}$ Figure of merit 1.0 Shunt Peaking case o

$$munt Feaking, case z$$

$$m = 0.41$$

$$C_t = C_1 + C_2$$

$$R_L = \frac{.85}{2\pi F_{max} C_t}$$

$$L_o = 0.41 R_L^2 C_t$$
Figure of merit 0.85

The shunt peaking circuit of Fig. 3B

Fig. 10. Time delay and amplitude characteristics of amplifier of Fig. 8.



is the simplest type as it involves only one inductance. It has the lowest figure of merit of any coupling network, but has the advantage of permitting relatively wide tolerances in circuit values with little change in performance. The design coefficient m relates the reactance of the peaking inductor L_{a} to that of the total shunt capacitance C_i at the frequency of correction, and determines the linearity of the phase characteristic up to the top frequency F_{max} . In case 2, improved phase response is obtained at the expense of gain factor. Figs. 4 and 5 show amplitude and time delay characteristics of the two shunt peaking networks, as a function of the ratio of actual to top frequency.

Four Terminal Networks

Four terminal networks give improved figures of merit, since the tube input and output capacitances are effectively isolated from each other. On the other hand, four terminal networks are more difficult to adjust, and frequently require certain specified ratios between the capacitances at each end. In general, both the figure of merit and the linearity of phase and amplitude response will be determined by the manner in which the capacitances are distributed; this is particularly true of the more complicated structures.

Case 1 Series Peaking (Fig. 6A).

$$C_{t} = C_{1} + C_{2}$$

$$C_{2} = 2C_{1}$$

$$R_{L} = \frac{1.5}{2\pi F_{max} C_{t}}$$

$$L_{e} = 0.67 R_{L}^{2} C_{t}$$
Figure of merit 1.5

The series peaking network of Fig. 6A provides a gain of 1.5 over the uncompensated case, at the same time permitting a moderate tolerance with respect to capacitance distribution and circuit values. It is sometimes possible to achieve a slightly greater figure of merit than the above figure would indicate by using higher values of inductance and load resistance, but this procedure may require resistive damping of L_c in order to obtain sufficiently uniform response.

It might be noted that the positions of the source and load may be interchanged so that R_L is at the output end, with no change in network characteristics. This may be found necessary in certain cases where the terminating capacitance C_2 is less than the input capacitance C_1 . In any case, the load resistor R_L should be on the low capacitance side of the peaking inductor L_c .

Capacitance ratios other than 2 to 1 or 0.5 to 1 may be accommodated at the expense of uniformity of phase and amplitude response. If, in Fig. 6A, the
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A

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INFRARED SPECTROPHOTOMETER

A double beam infrared spectrophotometer which will meet the demands of the structural chemist for high resolu-



tion and sensitivity, and the analytical chemist for speed and accuracy, is now in production at the plant of the *Perkin-Elmer Corporation*, Glenbrook, Conn.

Designated Model 21, the instrument records directly in percent transmission against a linear wavelength scale on large, easily read charts. The instrument's resolution is up to the limitations imposed by the Raleigh criterion and Johnson noise. Its speed of scanning ranges from 3 minutes to 100 hours for the rock salt region, and the wavelength drive speed can be suppressed automatically when desired. The over-all range of the instrument is from less than 2 microns to 15 microns in the rock salt region.

Chart scales of the *Perkin-Elmer* Model 21 are uniform from 1 to 50 inches per micron by integral factors. The amplifiers and power supplies are external.

PICTURE TUBE COMPONENTS

The Tube Department of *RCA*, Harrison, N. J., recently offered to equipment manufacturers five new components for the RCA 16GP4 deflection sys-



tems designed to use the horizontaldeflection amplifier tube 6CD6-G and the vertical-deflection amplifier tube 6S4.

These new components are as follows: Deflecting Yoke, Type 206D1; Width Control, Type 208R1; Horizontal-Linearity Control, Type 209R1; Horizontal-Deflection-Output and High-Voltage Transformer, Type 218T1; and Vertical-Deflection-Output Transformer, Type 222T1.

PARABOLIC ANTENNAS

Five new parabolic antennas to cover the 5929-7125 mc. frequency band are now available from the *Workshop Associates, Inc.*, 66 Needham St., Newton Highlands 62, Mass.

Each parabola is available in two, four, six, and eight foot diameters, and



mounts are available for all types of installations. The antennas have gains up to 44.9 db., and can be supplied with complete de-icing equipment and junction boxes.

A free booklet giving complete descriptions of these antennas may be obtained upon request.

TEST EQUIPMENT

A new instrument which indicates on the screen of an oscilloscope instantaneous speed vs. time is announced by *Kay Electric Co.*, Maple Avenue, Pine Brook, New Jersey. The Rotalyzer uses a properly magnetized disc and a pickup to provide a signal whose frequency is proportional to instantaneous speed. An accurate determination of this frequency is made and indicated on the cathode-ray oscilloscope screen. Average speed is read on a panel meter.

This instrument may be used in the study of torsional vibrations in rotating shafts, in the study of "wow" in phonograph equipment, and in the measure-

ENGINEERING DEPT.

ment of shaft rotational changes in computer systems. The r.p.m. range of 900 to 7200 r.p.m. may be extended down to 33-1/3 r.p.m. and upward to 50,000 r.p.m. Speed variation is up to 200 c.p.s.

ELECTRONIC TILE SORTER

An electronic tile sorter designed primarily to detect flaws in 4¼" x 4¼"



x $\frac{1}{4}$ " double-fired tile when it is in the bisque state has been announced by *Electronic Associates Incorporated*, Long Branch, N. J.

The Model 203 Tile Sorter is adjustable to meet any possible conditions of mixture or firing. Adjustment to any level desired by the tile manufacturer is dependent on the following ratios: 1) When set to reject 100% of the tiles with objectionable air pockets, it will also reject 100% of the cracked tile and will reject all tiles under 200 lbs. on the 3-point break test; 2) When set to reject 95% of the tiles with objectionable air pockets, it will reject 87% of the cracked tiles and will reduce the break test to 110 lbs; and 3) When set to reject 90% of the air pocket tiles, it will reject 80% of the cracked tiles and the break test is reduced to 80 lbs.

This sorter is capable of sorting tiles at any rate up to 10 tiles per second, which is much faster than the tiles can be brought to the machine using present conveyor methods.

HIGH SPEED RELAY

The Autocall Company, Shelby, Ohio, has released for general use their ultra high-speed "HHA" relays which were originally designed and built as a basic component of the Harvard University's



Mark II Calculator for the Navy Department.

These relays, according to the manufacturer, are built for long life with 100 million operations, minimum, at high speed of 6 to 8 milliseconds and incorporate many features including six single-pole double-throw contacts arranged side-by-side to provide 28 different contact combinations; individual contacts and prongs are of one-piece construction, and each contact provides its own prong for jack-connecting to the circuit.

Coils can be furnished for 24 to 150 volts, d.c., with no increase in price for the various voltages.

SOLDERING UNIT

A soldering unit, operating on the conduction principle, and suitable for all types of soldering operations including silver soldering and brazing, is now



available from Wasserlein Mfg. Co., Inc., 7400 3rd Ave., No., St. Petersburg 6, Florida.

The Wassco Glo-Melt has a 24 heat selector to handle lightest up to heaviest work. With capacity equal to a 450 watt heavy electric iron, Glo-Melt soldering is done with a light handpiece weighing only 5 oz., which is said to be as easily and accurately handled as a pencil. According to the manufacturer, this unit is faster and more accurate, consumes less power, and uses solder and flux.

Glo-Melt literature and a use-manual are available from the manufacturer.

TIME INDICATOR

Electronic Systems Co., 555 East Tremont Avenue, New York 57, N. Y., has announced its Model 632-B Pulse Rise Time Indicator for the accurate plotting of the rise time of rapidly rising positive voltage pulses. The device employs a specially designed delay line of variable length and a vacuum tube voltmeter.

The 632-B features a radical change in the method of analyzing the build-up time of a pulse. However, the most important factor is the range available: .005 to .1 microseconds in 20 steps. Other ranges are available upon request.

PILOT TUNNEL KILN

Harper Electric Furnace Corporation, Niagara Falls, N. Y., has announced a small tunnel kiln for speeding up pro-



duction test runs. Its versatile design permits variations in firing cycles to allow duplication of most firing schedules.

These units are being used in test firing of ceramic powders, steatite, electrical porcelain, insulators, resistors, spark plugs, grinding wheels, newly developed electronic components, and other ceramic products.

Further information may be obtained by writing the company direct.

FM SIGNAL GENERATOR

Boonton Radio Corporation, Boonton, N. J., has announced the Type 202-D frequency modulated signal generator designed for use with telemetering receiver equipment and in other associated applications.

This instrument covers the frequency range from 175 to 250 megacycles and is provided with three continuously adjustable deviation ranges: 0-24 kc., 0-80 kc., and 0-240 kc. In addition, amplitude modulation up to 50% may be obtained using the internal audio oscillator and



modulation to 100% using an external audio oscillator.

The deviation sensitivity of the frequency modulation system is within \pm 0.5 db. from d.c. to 200 kc. The ampli-(*Continued on page* 29)

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DR. LANGMUIR RETIRES

Dr. Irving Langmuir, regarded as one of the greatest scientists of modern



times, has retired as associate director of *General Electric Company's* Research Laboratory. Dr. Langmuir will continue work in the laboratory in a consulting capacity, but will engage primarily in activities of Project Cirrus, joint weather research program of the U. S. Army Signal Corps and the Office of Naval Research in consultation with *General Electric*.

Dr. Langmuir, who has received the world's top-ranking scientific awards, including the Nobel Prize in chemistry in 1932, was responsible for the development of the gas-filled incandescent lamp, the high-vacuum power tube, atomic hydrogen welding, screeningsmoke generator for the military, and methods for artificial production of snow and rain from clouds.

MICROWAVE REPEATER IN PRODUCTION

The *Philco* feedback-type microwave repeater announced at the recent annual session of the Communications Section of the Association of American Railroads is now being manufactured on a production basis.

Capable of handling up to 32 two-way voice channels or combinations of voice and coded intelligence, the *Philco* CLR-5 microwave repeater is designed for operation in the 6575-6875 megacycle band, which is available to railroads. This feedback repeater circuit permits multiple chain repeats with very little distortion and makes possible the use of only one microwave oscillator tube for both transmitting and receiving functions of a single-direction repeater.

The Industrial Division of *Philco Corporation*, Philadelphia 34, Pa., will give further information and detailed technical specifications upon request.

ELECTRIC FIELD

A shadow technique, developed by Dr. L. L. Marton and associates of the National Bureau of Standards, now makes it possible to photograph and study quantitatively electrostatic and magnetic fields of extremely small dimensions.

The striking pattern shown is the result of an electric field about a charged crystal of barium titanate when studied by the electron-optical shadow



method. This method offers a powerful tool for exploring fields that have not been susceptible to other methods of investigations.

An electron lens system is used to produce a shadow image of a fine wire mesh placed in the path of an electron beam. From the distortion in the shadow network caused by deflection of the electrons as they pass through the field under study, accurate values of field strength are computed. In the photograph, the distorted shadow image of the wire mesh is superposed on the image of the crystal (center) and its 0.010-inch tungsten-wire support.

NEW SITE FOR NBS LABORATORY

Approval has been given for the development of a site at Boulder, Colorado, for additional Bureau laboratory facilities. To be used initially by the Bureau's Central Radio Propagation Laboratory, the site consists of about 210 acres directly south of the city.

Laboratory facilities at Boulder for research in radio propagation at a cost of about \$4,500,000 have been authorized by Public Law 366 of the 81st Congress. Actual construction is expected to start in the summer of 1951 and when completed will maintain a research staff of about 300 people.

NBS TO USE MEMORY UNITS

The first of three *Technitrol* memory units has been obtained by the National Bureau of Standards for use in the latest type of large scale electronic computer developed jointly by funds of the Air Comptroller Department of the Army and NBS.

The memory unit, based on the principle of the mercury delay line originally conceived at the University of Pennsylvania and manufactured by the *Technitrol Engineering Co.*, is an automatic brain which will remember for later recall any sequence of facts or figures once placed on it. It has reached its latest development in the computer to be known as the NBS Interim Computer.

In addition to contemplated work in processing the 1950 census figures, the Interim Computer will be studied in an effort to develop new and better computers for Government and commercial use. It will also be utilized for solving many of the scientific problems that arise in the normal work of the Bureau.

ULTRASONIC GENERATOR

One of the many electronic developments described in "Electrical and Allied Developments of 1949" published in the *General Electric Review* was an ultrasonic generator having a quartz



transducer on top of the power supply and operating in an oil-filled well.

Power at 300, 500, 750, or 1000 kc. is supplied by the equipment. Applications of this ultrasonic generator include as-

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sistance to chemical reactions, bacteria control, and the mixing of immiscible liquids such as mercury into water.

IMPROVED MULTIPLIER PHOTOTUBE

Important advances in nuclear research, astronomy, photoelectric spectrometry, and other fields involving work with light at extremely low levels



is expected with the announcement of a greatly improved 1P21 multiplier phototube by *RCA's* Tube Department. Heart of the cyclotron used in radiation research at the University of Rochester is an *RCA* 1P21 multipler phototube shown partially disassembled. The "equivalent noise input" of the improved 1P21 has been reduced to 5×10^{-13} lumen at room temperature. This value represents a six-fold reduction in operational noise and permits a corresponding reduction in the lower limit of measurable light intensities.

This extension in the range of the 1P21 makes it an even more valuable aid for astronomers studying light from distant stars, for nuclear scientists studying atomic radiation, and for other laboratory research work requiring measurement of light of extremely low intensity.

INSTRUMENT TO RECORD BRAIN AND HEART ACTION

Equipment for making radar-like pictures of the heart and brain in action was described by Dr. Stanford Goldman, professor of electrical en-



gineering at Syracuse University and former research associate in the M.I.T. Research Laboratory of Electronics, at the recent IRE Convention. Results already obtained indicate that the pictures will be useful for the diagnosis of disease in the heart and brain and in studying the physiology of these vital organs.

The equipment, first conceived and developed at M.I.T., picks up tiny electrical impulses which flow through nerves in the body. These impulses are then converted into a constantly moving map-like picture similar to those made by World War II radar sets.

Equipment shown in the center of the photograph amplifies signals from electrical pick-ups on the patient at the right. Dr. Goldman at the left is shown examining a printed record of these signals while William F. Bantelmann, Jr., M.I.T. Research Assistant in electrical engineering, watches the visional pattern.

NEW LITERATURE

Handling of Radioactive Isotopes

The National Committee on Radiation Protection, established by the National Bureau of Standards, has prepared a handbook giving recommendations for the safe handling of artificially-produced radioactive isotopes in the typical laboratory or small industrial operation.

Handbook H42, Safe Handling of Radioactive Isotopes, gives general safety recommendations which may be modified to suit the control requirements of different types of operations. Specific problems discussed are those of personnel, laboratory design and equipment, hazard instrumentation, hazard monitoring, and transportation of isotopes. The increasing use of radioactive isotopes by industry, the medical profession and research laboratories has made it essential that certain minimal precautions be taken.

This handbook is priced at 15 cents a copy and is available from the Superintendent of Documents, U. S. Government Printing Office, Washington 25, D. C.

Paper Laminates

A report prepared by the Plastics Laboratory of Princeton University under joint sponsorship of the National Defense agencies describes the development of a low-loss, arc-resistant paper laminate for high frequency applications.

The report, released to the public by the Office of Technical Services of the U. S. Department of Commerce, deals with modifying standard melamine-formaldehyde impregnating resin with varying amounts of hexamethylenediamine to produce a laminated material with high dielectric strength,

(Continued on page 25)







DR. MERLE M. ANDREW has joined the staff of the machine development section of the National Bureau of Standards applied mathematics laboratories. Before coming to the Bureau, Dr. Andrew was engaged in operational research with the Naval Operations Evaluation Group and prior to that time conducted radar research at the Radiation Laboratory of MIT. He is a member of the American Mathematical Society and the American Physical Society.



ARTHUR L. CHAPMAN has been named general manager of the Colonial Radio and Television Division of Sylvania Electric Products Inc. With headquarters in Buffalo, N. Y., Mr. Chapman will be responsible for production and sales of the division in addition to his duties as general manager of Sylvania's newly formed Parts, Wire and Plastics Division at Warren, Pa. Mr. Chapman has been associated with Sylvania since 1933.



CURTIS R. HAMMOND has been appointed equipment sales manager of the Receiving Tube Division of *Raytheon Manufacturing Company*. He will be responsible for the sale of radio receiving tubes and cathode-ray picture tubes and will direct activities from *Raytheon's* Chicago warehouse. Mr. Hammond joined *Raytheon* five years ago after many years in sales and sales engineering work with *Ken-Rad Tube and Lamp Corporation*.



DELMAN E. ROWE has been appointed senior test director in the guided missiles research laboratory of the National Bureau of Standards. Mr. Rowe, who received the Naval Ordnance Development Award with Certificate by the Naval Bureau of Ordnance in 1945, is a member of the Institute of Radio Engineers, is the inventor of a microwave switch and attenuator, and has published several technical papers in the electronics field.



ROBERT A. STAREK has been appointed field engineer for the Radio Tube Division of *Sylvania Electric Products* at Emporium, Pa. Formerly commercial engineer, Mr. Starek joined the Engineering staff of *Sylvania* immediately after receiving his B.S. degree in electrical engineering from Iowa State College. He is a member of the American Institute of Electrical Engineers and the IRE and will make his headquarters in the Cincinnati office.



JEROME R. STEEN, director of quality control for Sylvania Electric Products, Inc., New York, has been elected to grade of Fellow by the board of directors of the Institute of Radio Engineers. Mr. Steen joined the engineering staff of the Radio Division of Sylvania at Emporium, Pa., in 1931 and since that time has been active in the application of quality control and standardization methods in radio tube manufacture.

New R-C Osc.

(Continued from page 14)

plate load resistor R_{L2} or the alternate cathode resistor R_{k1} may be varied for manual adjustment of amplifier gain. The higher supply voltages give a more linear operating region for the amplifier tubes.

Oscillators of this type have been adjusted to operate with various supply voltages ranging from 100 to 350 volts. When distortion is considered important the circuit should be supplied with higher voltages and adjusted to as near the oscillation threshold as is deemed advisable. The supply should be regulated if possible but satisfactory results can be obtained over quite a range of voltages.

The amplitude of oscillation is a marked function of these supply voltages; therefore, the relative distortion will also vary. Some compensation is possible by using one of the circuit variations.

Circuit Variations

There will be some quiescent current flowing in the nonlinear element if two equal resistors are used in the cathodes. These tubes are not balanced because the load resistor appears only in the plate circuit of stage two. This condition will be known as the unbalanced circuit.

If the cathode resistor R_{k2} is increased in value the d.c. voltages on the cathodes may be made equal and the nonlinear element will not carry d.c. current. This will be known as balanced circuit A and does not require any additional circuit components.

When the circuit is made symmetrical by using a dummy load resistor in the plate circuit of the first stage the quiescent current may be removed from the nonlinear element. This will be known as balanced circuit B. If this dummy load is bypassed for a.c. signals the gain is given by the first expression. If the dummy load is not bypassed the element R_{L1} must be included in calculation of the net amplifier gain in this manner:

$$K' = \frac{A}{B + CR_z} \qquad (9)$$

where the constant A is the same as Eqt. (6), and B and C are given by:

$$B = (r_{p} + R_{L2}) [r_{p} (R_{k1} + R_{k2}) + (\mu + 1) R_{k1} R_{k2}] + r_{p} R_{k1} R_{k2} (\mu_{2} + 1) + R_{L1} \{ R_{k2} (r_{p} + R_{L2}) + R_{k1} [r_{p} + R_{L2} + ... (10) \\ (\mu_{2} + 1) R_{k2}] \} ... (10)$$

$$C = [r_{p} + (\mu_{1} + 1) R_{k1}] [r_{p} + R_{L2} + ... (10)$$

 $\begin{array}{c} (\mu_{2}+1)R_{k2}] + R_{L1} \left[r_{p} + R_{L2} + \\ (\mu_{2}+1)R_{k2} \right] \end{array}$ (11)

The basic unbalanced circuit causes the attenuation in the " π coupling" network to increase with increasing supply voltages. This tends to counteract

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A 6SN7 oscillator of the unbalanced circuit form operating at 995 to 997 c.p.s.

the increase in tube amplification at higher supply voltages in much the same manner as the circuit of Fig. 7A. This unbalanced circuit offers this inherent compensation; however, where the greatest a.c. sensitivity is desired the amplifier should be balanced for equal d.c. voltages on the cathodes.

The variation of amplifier gain with respect to signal level has been examined for the unbalanced and two balanced circuit variations. Typical results are indicated in Fig. 8.

The basic unbalanced circuit is shown to be less responsive to supply voltage variations in the results given in Fig. 9.

The circuit presented is well suited to fixed frequency oscillators. The values required may first be calculated and then trimmed on a trial and error basis. A great deal of care is required to locate the frequency to less than 1 per-cent using fixed elements. These rules may prove helpful:

1—To raise the frequency pad additional resistance across R_1 or R_2 in the feedback circuit.

2—To lower the frequency pad additional capacitance across C_1 or C_2 in the feedback circuit.

3—The choice between the two resistors (or the two condensers) is made considering the present level of oscillation. When additional resistance is placed across R_1 or additional capacitance is placed across C_1 the signal level of steady oscillation will increase. If some adjustment of amplifier gain is provided this need not be considered.

A number of fixed frequencies may be obtained by suitable switching of fixed components in the feedback network. Both the resistance and the capacitance elements may be switched for frequency ranges greater than one decade. There is a relationship between the values selected; these general rules may help in adjusting these values. 1—Begin with the larger values of resistance and capacitance first.

2—Complete the selection of the set of elements (R's or C's) which are greater in number. Attempt to maintain nearly constant output signal amplitude for these values.

3—Complete the selection of the remaining set of elements. This set will generally be used as multipliers of the frequencies given by the first set.

A very useful audio oscillator may be constructed using two sets of element values for the feedback network. An example of the different frequencies which might be obtained is as follows:

The resistor elements are selected with a two pole, nine point switch to give these base frequencies: 10, 15, 20, 25, 30, 40, 50, 60 and 80 cycles per second. The condenser elements are selected with a two pole, three point switch to give these multiplier factors: Times 1, Times 10 and Times 100.

Continuously variable frequency oscillators may use ganged resistance or capacitance pairs with the alternate elements switched for different multiplier factors times the basic frequency range. Ganged resistors may be used in series with fixed resistors to minimize tracking errors-a frequency range of 3:1 may be suggested here. Variable resistor elements will cause the loading of the feedback network upon the amplifier to vary as the oscillator is tuned to different frequencies. Ganged condensers may be used for a range of 10:1 as tracking is much easier and the loading upon the amplifier is constant. If the minimum capacitance is to be 100 $\mu\mu$ fd. the condenser pair would require a maximum of 1000 $\mu\mu$ fd. per section for a decade of frequency variation. With care a smaller minimum capacitance might be used.

Advantages of this new oscillator are primarily in design and economics,



Fig. 13. Gain characteristics of the amplifier shown in Fig. 7A.

where cost, weight and space are important. Fixed frequency applications include bridge test sets, intermodulation test sets, modulation signals for signal generators, telemetering systems, tone generators, and simple musical instruments.

Semi-variable and continuously variable frequency oscillator applications are many and varied. This oscillator will best serve in those applications where it has not been possible to use other circuits because of the cost, space and weight factors.

News Briefs

(Continued from page 23)

very low dielectric loss and good punching characteristics.

PB 99220, Low-loss Arc-resistant Paper Laminates, sells for 50¢ a copy. Orders should be addressed to the Office of Technical Services, U. S. Department of Commerce, Washington 25, D. C., and should be accompanied by check or money order payable to the Treasurer of the United States.

Welding Control Equipment

Westinghouse Electric Corporation now has a booklet describing in detail its electronic resistance welding control equipment, both synchronous and nonsynchronous.

Basic equipment in the control circuit includes a sequence-weld-timer panel and a means to fire the ignition tubes in the power circuit. Auxiliary control panels can be readily added to meet specific job requirements.

A copy of this booklet, B-4309, may be obtained by writing *Westinghouse Electric Corporation*, P. O. Box 2099, Pittsburgh 30, Pa.

TV Tube Booklet

The Radio Tube Division of Sylvania Electric Products Inc., Emporium, Pa., is offering a new 20-page booklet providing television picture tube and general purpose cathode ray tube characteristics; replacement tube data; base diagrams; suggestions for tube handling; and a concise description of cathode-ray oscilloscope use in television servicing.

Information contained covers 165 tube types with faces ranging from two to twenty inches maximum dimension utilizing electrostatic or magnetic deflection systems. A nomenclature chart explains the meaning of type number letters and figures and applications for different types of tube screens.

The booklet is available through Authorized Sylvania Distributors.

Geiger-Muller Counter

The nature, construction, and use of the Geiger-Muller Counter is presented in a new booklet just published by the National Bureau of Standards.

In addition to the treatment of the counter itself, the booklet discusses methods of detecting counter pulses, applications of counters to quantitative measurements, proportional counters, and the preparation and filling of Geiger-Muller Counters.

Circular 490, The Geiger-Muller Counter, may be obtained from the Superintendent of Documents, U. S. Government Printing Office, Washington 25, D. C., 20¢ a copy.



SYLVANIA TUBES

Subminiature Tubes

Two additional Sylvania subminiature tube types, a medium-mu triode



and a high-mu triode, are now available. Type 5645 medium-mu triode is a T-2 suitable for Class A amplifier applications. Under typical operating conditions the tube will have a transconductance of 2700 micromhos and an amplification factor of 20. Maximum rated plate dissipation is 1 watt and plate resistance is 7400 ohms.

Type 5646 high-mu triode is a T-2 suitable for Class A amplifier or resistance coupled amplifier applications. Under typical operating conditions the tube will have a transconductance of 2400 micromhos, an amplification factor of 70, and a plate resistance of 29,000 ohms. Maximum rated plate dissipation is 0.3 watt.

Both subminiature types have 6.3 volt, 150 milliampere heaters, and flexible leads for direct wiring to circuit.

Beam Power Amplifier

A high-perveance beam power amplifier designed for use as a horizontal



deflection amplifier in high efficiency deflection circuits for television receivers is another of the many new tubes announced by Sylvania.

Features of type 6AU5GT include: low-mu, high plate current at low plate voltage and high operating ratio of plate current to number 2 grid current. The tube has a 6.3 volt, 1.25 ampere heater. Under typical operating conditions, it will have trans-conductance of approximately 6000 micromhos; mu approximately 5.9. The maximum plate dissipation is 8 watts; peak positive pulse plate voltage is 4500 volts; and maximum d.c. plate voltage is 450 volts.

Picture Tubes

An 8¹/₂-inch, all-glass, direct-view television tube with electrostatic focus and deflection for use in television sets designed for 7-inch viewing tubes is now available through authorized *Sylvania* distributors.

The tube, type 8BP4, is interchangeable with type 7JP4 and offers the advantages of 50% increase in useful area. Deflection sensitivity is provided for full scan in circuits designed for full scan



with 7-inch tube types. Operating voltages include: 6000 on number two anode; 1620 to 2400 on number one anode; and zero to -72 to -168 volts on number one grid.

Type 16LP4 is a 16-inch, all-glass, direct-view television tube supplied with an external conductive coating which acts as a filter capacitance when grounded. The tube is supplied with neutral or clear face plates and employs an ion trap gun for use with external magnet.

Typical 16LP4 picture tube operating conditions include: maximum usable face diameter, 15% inches; heater volts, 6.3; heater current, 0.6 amperes; focus coil current, 110 milliamperes; ion trap magnet current, 120 milliamperes; number 1 grid volts (for cut-off) -33 to -77; number 2 grid volts, 300; and anode volts, 12,000.

Miniature Portable Radio Tubes

A new line of miniature radio tubes for portable radio receivers, said to be the first announced in this country in more than ten years in which filament current has been reduced below 40 milliamperes per tube, is now being offered by *Sylvania*.

The new tubes require only 25 milliamperes per tube which means that "A"



batteries used in portable receivers will last approximately three times as long as with previous tubes. All are supplied with 7-pin miniature button bases.

Types included are: 1U6, a heptode converter with oscillator anode as a separate element; type 1AF5, a diode pentode; type 1AF4, a sharp cutoff r.f. pentode; and type 3E5, a beam power output tube. Power required for a complement of the new tubes in a typical battery-operated superheterodyne is only 2.1 watts. They will also operate satisfactorily over a range of 1.4 volts to 1.1 volts.

Miniature Rectifier Tubes

The Radio Tube Division, Sylvania Electric Products Inc., 500 Fifth Ave., New York 18, N. Y., has announced a miniature high voltage half wave rectifier designed for television receiver pulse rectifying systems and voltage doubler circuits for magnetically deflected 10 and 12-inch viewing tubes.

The tube, type 1V2, has a peak inverse plate voltage of 7500, a peak plate current of 10 milliamperes, and an av-



erage plate current of .5 milliamperes. Type 1X2 is a double ended miniature high voltage rectifier tube designed for use with r.f., fly-back, and 60-cycle types of power supply for television picture tube anodes. This type is for use in power supplies where voltages up to

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15,000 volts d.c. are required. The 1X2 is mounted in a T-6½ bulb and has an over-all height of 2^{11} /₁₆ inches.

RECEIVING TUBES

Three new receiving tubes designed mainly for television receivers have been added to *General Electric's* tube production lines.

The 6AS5 is a beam-power amplifier of miniature construction intended for use as the audio power-output tube in television receivers and small radio receivers. When operating Class A1, with



a plate voltage of 150 volts and an input signal of 8.5 volts peak, 2.2 watts of output power can be realized with 10 per-cent distortion.

The 6BA6-GT and 25BQ6-GT are beam-power amplifier tubes designed to withstand high-surge plate voltages for short periods of time. These tubes are intended for use as horizontal-deflection amplifiers in television receivers. Maximum ratings of the tubes include a plate dissipation of 10.9 watts, a plate current of 100 milliamperes, and a peak positive surge plate voltage of 5000 volts.

Further information may be obtained from the Tube Divisions, Schenectady, New York.

RCA TUBES

Deflection Amplifier Tubes

Two new deflection amplifier tubes designed particularly for use with the



new 16-inch "metal" picture tube 16GP4 have been announced by the Tube Department of *RCA*, Harrison, N. J. The 6CD6-G shown is a high-perve-(Continued on page 29)

Video Amp. Design

(Continued from page 18)

ratio C_2/C_1 is designated by M, the values of R_L and L_c will be modified by:

$$R_L' = \sqrt{MR_L/2}$$
; $L_c' = ML_c/2$

where R_L' and L_e' are the new values of load resistance and peaking inductance respectively. The curves of Fig. 7 show amplitude and time delay characteristics of the series peaking network, for the case where $C_2 = 2C_1$.

Two Section Filter Coupling

It is frequently possible to secure uniform response and an improved figure of merit with a four terminal coupling network designed on the basis of a low-pass filter. Although such a network requires rather carefully controlled distribution of capacitances for best results, it does provide a gain factor approaching 2 over the uncompensated case, and has the further advantage that two output terminals are available (Fig. 6B).

The filter structure consists of a Ttype constant K full section, with midshunt terminated half-sections at each end. The full T section is composed of C_2 and the two series impedances $L_1/2$ and $L_2/2$. The load resistance and the cutoff frequency of the filter as a whole is determined by the capacitance C_2 . The cut-off frequencies of the end half-sections are determined by capacitances C_1 and C_3 . In order to prevent an unwanted rise in the image impedance of the input half-section, it is usually necessary to add resistive damping across L_1 . The value of the damping resistor R_d is best determined by experiment, and will generally lie in the vicinity of 5 to 6 times the load resistance.

Due to the larger number of reactive elements, this type of coupling network shows a somewhat greater deterioration of phase characteristics about the top frequency. In order, therefore, to obtain sufficiently uniform response, it is sometimes necessary to design for a higher top frequency than would otherwise be the case. For optimum results, the terminating capacitances C_1 and C_3 should be nearly equal to one half the center capacitance C_2 .

 $C_{t} = C_{1} + C_{2} + C_{3} \quad ; \quad C_{2} \simeq 2C_{1} \text{ or } 2C_{3}$ $R_{1} = 1/\pi F_{max}C_{t}$ $L_{1} = \left(\frac{1}{2} + \frac{C_{1}}{C_{2}}\right) \quad \left(\frac{R_{1}}{2\pi F_{max}}\right)$ $L_{2} = \left(\frac{1}{2} + \frac{C_{3}}{C_{2}}\right) \quad \left(\frac{R_{1}}{2\pi F_{max}}\right)$ Figure of merit $\simeq 2.0$

Case 3. Modified Shunt-Series Peaking

This type of coupling network provides a large figure of merit and at the same time quite uniform phase and amplitude characteristics. It is similar to

E N G I N E E R I N G

the two section filter coupling described above, except that the terminating halfsection is of the mid-series type. C_3 of Fig. 10 is therefore missing. The shuntseries network may afford a slightly better gain factor than the network of Fig. 6B, and the deterioration of phase response about the top frequency is not so great (Fig. 9A).

The accompanying Table II, taken from a report by O. H. Schade of RCALaboratories gives the information necessary to design a shunt-series peaking network for capacitance ratios of 0.3, 0.5, and 1.0. Since the characteristics of the network are comparatively insensitive to this ratio, it is possible to interpolate between the values given with a fair assurance of success.



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Of the six circuit values in the network, only two need be determined in order to completely design the interstage for a given bandwidth or gain. As the capacitances C_1 and C_2 are presumed fixed and reduced to their practical minimums, either the bandwidth or the gain may be used as a design parameter. Once the capacitances have been evaluated, the remaining parameters may be determined as a function of the chosen parameter directly from the table. The necessary design equations are:

$$C_t = C_1 + C_2$$
; $R_0 = 1/\pi F_c C_t$
 $L_1 = R_1^2 C_t$; $L_2 = 0.5 L_1$

where F_c is the cutoff frequency of the network and R_c is the load resistance design constant.

The general procedure in designing a network of this type is to determine first the capacitances C_1 and C_2 , the capacitance ratio C_1/C_2 and the total capacitance C_t . The ratio of the network cut-off frequency F_c to the top frequency of correction F_{max} for the existing capacitance ratio is determined from Table II (col. 4). The resulting value of F_c is then used to calculate the load resistance design constant R_c . The actual load resistance R_L can then be obtained from column 3 and the values of L_1 , L_2 , and R_d determined.

Where the video amplifier provides a fairly large gain reserve, it is sometimes possible to make use of selective degenerative feedback to compensate for discrepancies in the over-all high frequency characteristic. One simple method of accomplishing this is shown in Fig. 9B.

The resistance-capacitance network in the cathode circuit of the tube has a short time constant, and provides selective degeneration. C_k is a small mica capacitor, seldom larger than 1000 $\mu\mu$ fd.

The voltage gain of the stage at low and medium frequencies where C_k is not effective is given by:

$$\overline{A_L} = \frac{G_m R_L}{1 + G_m R_k}$$

and is, of course, always smaller than the stage gain were R_k omitted. The gain reduction due to R_k is:

20 log $(1 + G_m R_k)$ decibels

This equation indicates the maximum degree of amplitude compensation ob-

Table II.	Table for determining the
ratio of th	e network cut-off frequency
to the to	p frequency of correction
for the e	existing capacitance ratio.

1	2	3	4	5
C_1	R_{d}	R_1	$F_{max} F_{t}$	igure
C_2	\overline{R}_{1}	$\overline{R}_{ m o}$	F. a	of
			1	16/11
0.3	5.66	1.088	0.96	2.1
0.5	5.66	1.075	0.85	1.8
1.0	18.90	1.00	0.76	1.5

Tube Type	G_m	F'	F''
6AG5	5000	98	232
6AH6	9000	117	238
6AK5	5100	119	242
6AU6	5200	79	157
6AC7	9000	90	184
6AG7	11000	85	177
7B5 (6K6GT)	2300	24	47
7C5 (6V6GT)	4100	35	71
6L6G	6000	40	84

Table	I.	The q	gain-be	andwidth	factor
for so	me	of the	more	common	tubes.

tainable with cathode peaking for a fixed value of R_k .

At the high frequency end of the video band, the shunting reactance of C_k causes a decrease in the effective cathode circuit impedance, and results in an increased voltage gain. At the same time, the a.c. plate-cathode voltage is shifted in phase in a direction such as to partially counteract the phase shift produced by the plate circuit time constant $C_t R_L$. The gain at high frequencies is given by:

$$A_{h} = \frac{G_{m} R_{L}}{1 + (G_{m} R_{k}/y)}$$

where $y = [(2\pi f C_k R_k)^2 + 1]^{\frac{1}{2}}$ and the added phase shift by:

$$-\tan^{\cdot 1}\left(\frac{1}{2\pi f \ C_k R_k}\right)$$

It is possible to effect nearly complete high frequency phase and amplitude correction with cathode peaking by making the plate circuit and cathode circuit time constants equal, that is:

$$R_L C_t = R_k C_k$$

The figure of merit of a stage using cathode peaking only is generally not nearly so good as will be obtained with the two and four terminal networks described. Cathode peaking is, however, useful in small and controlled amounts to supplement compensation obtained by other means.

The high frequency characteristics of a well designed television video amplifier should be such that the maximum variation in over-all delay time at 4.0 megacycles will not exceed about 0.08 microseconds. This requires that the total phase deviation up to the highest frequency of importance be not greater than 57° per stage in a two stage amplifier. The over-all amplitude characteristic should be flat to within ± 1 decibel or so up to the top frequency of correction.

The schematic diagram of Fig. 8 shows a typical two stage television video channel incorporating high and low frequency compensation, while Fig. 10 indicates its time delay and amplitude characteristics.

"ACOUSTICAL DESIGNING IN ARCHITECTURE" by Vern O. Knudsen and Cyril M. Harris. Published by John Wiley & Sons, Inc., 440 Fourth Ave., New York 16, N. Y. 457 pages. \$7.50.

Architects, builders, acoustical engineers, teachers and students of architectural acoustics will find this book extremely valuable as it covers the entire field of acoustical design in architecture. Practical applications are emphasized and includes design of auditoriums, theaters, school buildings, homes, churches, radio, television and sound-recording studios.

The contents is divided into two sections: (1) a consideration of the general principles and procedures which form \mathbf{a} basis for all acoustical designing, and (2) a description of specific applications of these principles and procedures.

Complicated mathematical formulas are translated into physical explanations, and charts and nomographs are given for many formulas. Comprehensive tables give pertinent data on soundabsorptive materials and sound-insulative structures.

This volume will serve as a practical guide to good acoustical designing as well as a handbook in the solution of problems encountered.

"ADVANCES IN ELECTRONICS"

Volume II, Edited by L. Marton, National Bureau of Standards. Published by *Academic Press Inc.*, 125 East 23rd St., New York, N. Y. 378 pages. \$7.60.

Because of the favorable reception given the first volume of "Advances in Electronics," it was decided that a yearbook of this kind would receive wide acclaim. This second volume covering the latest advances in the field of engineering electronics includes a large contribution from European authors, whereas the first volume was written in its entirety by American authors.

Contributors to this second volume are: Donald K. Coles, Westinghouse Research Labs; H. Fröhlich, University of Liverpool, G. F. J. Garlick, University of Paris, France; Gunnar Hok, University of Michigan; G. Liebmann, Associated Electrical Industries, Ltd., Berkshire, England; Hilary Moss, Electronic Tubes Ltd., Reading, England; George T. Rado, Naval Research Laboratory; and J. H. Simpson, National Research Council of Canada.

The editors and publishers have expressed a hope that this text will be a definite contribution to the free and international exchange of ideas among scientists.

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New Products

(Continued from page 21)

tude modulation system is substantially flat from 30 cycles to well above 100 kc. A monitoring meter is used to standardize the output level of the signal generator in order to make the mutual inductance r.f. attenuator direct reading over the range from 0.1 microvolts to 0.2 volts.

The instrument is self-contained, with power supply, and is designed for use on 115 volts, 60 cycles. Complete details may be obtained on request.

COIL WINDER

A manually-operated coil winding machine designed for the production of paper-insulated coils in multiple or "stick" form is announced by Universal Winding Company, Providence, Rhode Island.

The No. 108 Winder was designed to supplement the automatic type of machine with a low-cost machine suitable for the predominantly "job shop" type of market characteristic of the everchanging electrical and electronic parts industries. The machine can be adjusted quickly to accommodate changing requirements of wire size, coil length and diameter.

Power is supplied by a $\frac{1}{2}$ h.p. a.c. constant speed motor driving through an adjustable-sheave speed controller to a multiple-disc friction clutch attached directly to the spindle. Speed range is 400 to 2200 r.p.m.

VOLTMETER

A new model of the Mini-Volt Voltmeter manufactured by *Industrial De*vices, *Inc.*, Edgewater, N. J., which features an expanded scale centered on the common 110 and 220 line voltages, is announced.

The new adaptation of this voltmeter is known as Model 410A and is accurate to within 2 volts at 110 volts a.c. Practically burnout proof operation is assured by the glow-lamp indicator which is guaranteed for 25,000 hours' operation minimum. 12" flexible test leads are tipped with heavily insulated test prods assuring user maximum safety.

The Mini-Volt can check voltage, a.c. or d.c.; check for continuity; check blown fuses; locate grounded components; warn of live wires and scores of other possible troublesome occurrences known to electricians, refrigeration servicemen, maintenance men, and anyone who works with electrical circuits.

PRECISION POTENTIOMETERS

Technology Instrument Corporation, 1058 Main St., Waltham, Mass., is now manufacturing an improved version of their Type RV3 precision potentiometers. Available in two types, the Type RV3-8 (8 watts) and the Type RV3-12 (12 watts), these potentiometers have been redesigned to include bronze bushings for the rotor shaft, tapped mounting inserts, rotor take-off slip ring, dustproof construction with Bakelite cover fastened with screws, and molded parts of low-loss mica-filled Bakelite.

The Type RV3 potentiometers can be supplied with a wide variety of nonstandard features to meet special requirements, and the manufacturer has published an established price scale covering such non-standard features. Complete information may be obtained by writing the company direct.

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New Tubes

(Continued from page 27)

ance, beam power amplifier featuring low mu-factor, high plate current at low plate voltage, and a high operating ratio of plate current to grid No. 2 current. In suitable horizontal-deflecting circuits, only one 6CD6-G is required to deflect fully any directly viewed kinescope having a deflection angle up to 70° and operating at an anode voltage up to 14 kilovolts.

The 6S4 is a high-perveance, mediummu triode of the 9-pin miniature type. In suitable vertical-deflecting circuits, the 6S4 will deflect fully a 16GP4 or any other similar kinescope having a deflection angle up to 70° and operating at an anode voltage up to 14 kilovolts.

Sharp-Cutoff Pentode

The 6AS6 announced by RCA is a sharp-cutoff pentode of the 7-pin miniature type. Designed so that grid No. 1



and grid No. 3 can each be used as independent control electrodes, the 6AS6 is especially useful in gated amplifier circuits, delay circuits, gain-controlled amplifiers, and mixer circuits.

This tube is also suitable for use as an r.f. amplifier at frequencies up to about 400 megacycles per second. $\neg \otimes \neg$

Surface Wave

(Continued from page 11)

several months and had a rather thick corrosion layer. In this state the measured loss was 3 db. Most of the corrosion was then removed by sandpaper. The result was an increase of the loss to about 3.6 db. The expected loss for a wire with perfectly clean surface is much higher. Then we applied several coats of polystyrene solution by means of a brush. The loss went down more and more till a minimum of 1.7 db. was reached. Further increase of the dielectric layer gave a slight increase of the loss by about 0.2 db. The thickness of the dielectric coat varied more than 1:10 along the wire, thus it was not possible to measure the average thickness. This test indicated that the wave is not sensitive to inhomogeneities of the guide.

Finally, a test was conducted with a 600 ft. line which showed that the support of the wire is not a serious problem. The horns used were the same as in the other experiments (opening of 13"). The wire was an enameled wire of 3.2 mm. thickness. The enamel coat was 25/100 mm. thick and the frequency was 1600 mc. The line was stretched along the slope of a hill with the distance from the ground varying between 4 and 8 feet. The wire was supported at intervals of about 80 feet by waxed strings. The calculated loss was 4.5 db. and the measured loss 5 db. For comparison, an RG-8/U cable of the same length has an attenuation of 70 db. The supports and the bends in the wire produced by the supports had no measurable effect in the attentuation.

Fig. 4 shows the calculated attentuation in db. per mile for a line with 2 cm. diameter and a polystyrene coat of $\frac{1}{2}$ mm. in the frequency range from 300 to 1000 mc. In addition to this loss which is proportional to the length we get the loss of the launching devices, which could be kept below 2 db. Considering the wide frequency band which could be transmitted by such a line, the loss is very small.

The question of the effect of weather conditions can be answered only partially and only as far as rain is concerned. An increase of the attenuation was observed only if the wire was covvered with rows of big raindrops. The maximum increase measured so far for 1600 mc. was less than 1.5 db. A water film did not increase attentuation. It enlarges the thickness of the dielectric layer and behaves like a good dielectric because the ratio of the power factor to $(\varepsilon - 1)$ which determines the losses is very small. The effect of snow and ice will certainly be more serious and it may be necessary to prevent the settling

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down of snow by shaking the wire or by electric heating.

It is a pleasure to acknowledge the close cooperation in Coles Signal Laboratory which made it possible to obtain the present results in a relatively short time. Thanks are due to Mr. J. Hessel and R. Lacy, Chief and Assistant Chief, Radio Communication Branch, to Mr. J. Egli, Chief, Radio Relay & Microwave Section, and especially to the group who cooperated in the measurements, Mr. C. Sharp and L. Battersby and particularly Mr. A. Meyerhoff, who checked all calculations and improved the English of the paper which otherwise would have been much worse.

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Microwave Trans.

(Continued from page 9)

where T is the time required for one period (shown in Fig. 9).

The integral is merely equal to the pulse area. For the rectangular pulse shown in Fig. 9, this area is equal to the amplitude times the pulse width. The relation between average power and peak power for a rectangular pulse is therefore equal to:

Thus it is seen that if the period is equal to 1000 microseconds, and the pulse width equal to 1 microsecond, the peak power is equal to 1000 times the average power.

In a lighthouse tube the maximum power rating, for a given frequency, is primarily limited by the maximum allowable plate temperature or the average power dissipated. As a result, a peak power thousands of times greater than the maximum average power rating of the tube can be effected. Fig. 6 is a series of typical power output versus frequency curves for different plate voltages for a 2C43 tube modulated by the pulse series shown in Fig. 9.

In pulse modulating systems where the timing of the pulse carries the intelligence an effect known as "jitter" may be introduced into the microwave oscillator. The r.f. oscillator is normally keyed "on" at some time during the rise time of the applied pulse - the exact instant of keying depends upon the tube characteristics, circuit constants, and random noise fluctuations. If all of these conditions remained the same, the r.f. oscillation would start at the same relative instant for each succeeding pulse. However, the random fluctuations of the noise signal may cause the tube to fire either slightly before or after the normal repetition rate. Since the timing of the pulse carries the intelligence, this shift in pulse position appears as noise or "jitter" in the demodulator of the system.

This effect can be minimized through the use of a "catalyzer".6, 7 A "catalyzer", as many readers may remember from their chemistry courses, is an element which aids a reaction but does not react itself. In this case the catalyzer consists of a small oscillator which injects a signal of approximately the same frequency as that of the r.f. oscillator into the microwave cavity. The catalyzer oscillator can be a conventional pushpull oscillator operating in the 300 to 500 megacycle band with a crystal multiplier used to obtain the fifth or sixth harmonic of this signal. The output of the crystal multiplier is then fed to the cavity through a probe or loop. It has been found that the use of a catalyzer appreciably improves the signal to noise ratio of the microwave oscillator. The reason for this effect is probably that the "catalyzer" signal is sufficiently large to "overpower" the noise. Thus random changes in noise have little effect on the oscillator.

There are a number of parameters that affect the frequency of a lighthouse oscillator. Among these effects are the change in cavity length due to a variation in temperature (as a result of expansion and contraction of conductors), change in plate or filament voltage (affecting electron transit time), and cathode to grid interelectrode capacitance (feedback effects). In a typical commercial transmitter, it has been found that the cumulation of all these effects (using "self compensated" temperature conductors in the coaxial lines) results in a frequency stability of about 0.05 per-cent at 2000 mc. In this latter unit no attempt had been made to compensate for the voltage variations through the use of a regulated plate supply.

Where a higher degree of frequency stability is required, some method of automatic frequency control (a.f.c.) must be employed. Heretofore it has been found that the best method of frequency control is an electromechanical system in which a variation in frequency from the desired value is detected by a discriminator, whose output actuates a motor which drives a tuning device until the desired frequency is

Fig. 10. Frequency response of two reference cavities and desired resonant frequency of triode resonant line.



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Fig. 9. Typical pulse series.

obtained. Completely electronic systems of control are difficult and expensive to design since the easiest and surest method of tuning involves adjustment of the resonant line length, rather than variation of plate voltage or interelectrode capacitance, which is essentially a mechanical operation.

Two typical a.f.c. circuits are shown in Figs. 5 and 8. The a.f.c. circuit shown in Fig. 5 uses two reference cavities, one resonant at f_1 and the other at f_2 . The desired resonant frequency of the triode line, f_0 , is equal to $(f_1 + f_2)/2$. These relations are plotted in Fig. 10.

A small amount of energy from the triode line is picked up by loops shown in Fig. 5 and injected into the two reference cavities. When the frequency of this energy is exactly equal to f_o , the outputs of the two cavities are equal and no discriminator voltage is developed across the grids of V_1 and V_2 . V_1 and V_2 are thyratrons which are normally cut off. If the energy obtained from the triode line is not equal to f_o , a discriminator voltage will be developed across the grid of V_1 or V_2 (depending upon whether frequency is above or below f_o) causing one to conduct and operate the split phase motor. This motor will operate the reset control in the direction that will correct for the "off frequency" variation. An a.c. voltage is applied to the plates of the thyratrons, so that they will be extinguished when the discriminator voltage reduces to zero. The cavities used can be made highly reliable by using temperature compensated metals such as Invar or using a temperature controlled oven.

The a.f.c. circuit shown in Fig. 8 is a simplified version of the previous circuit. In this case a single cavity and thyratron are used in conjunction with a motor driven trimmer condenser. The effect of a condenser at the end of a line was discussed in the article, "Microwave Components". The action of this circuit is as follows: The existence of a frequency other than the desired value in the cavity provides a discriminator voltage which fires the thyratron. This operates the motor which drives the condenser rotor. The plate of the condenser turns very slowly, first increasing and then decreasing resonant line frequency. The plate will turn until the discriminator voltage drops to zero. The motor speed is made sufficiently low compared to the time required for the discriminator to extinguish the thyra-

tron, that the condenser will not "overshoot" the desired frequency.

Acknowledgement is hereby gratefully made to Mr. Harris Gallay of Federal Telecommunication Laboratories for automatic frequency control circuits described in this article.

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~®~

Custom Sound

(Continued from page 5)

finely, but it must be remembered that at certain times there were high levels of noise to overcome on stage due to the locomotives, western style gunplay, and old one-lung autos. If the system were not adjusted so carefully, much of the dialogue and music would be lost in some areas while in other areas the level of the sound would be so high as to be objectionable.

An interesting note is the fact that these speakers and baffles were so efficient in projecting the sound to the grandstand that additional monitor speakers had to be installed backstage in the wings to allow the stage managers and cast to follow the continuity. These were Jensen eight inch heavy-duty loudspeakers mounted in weatherproof metal projectors. Each of these monitors had an adjustable "L" pad control to allow the stage managers to set the level to suit their surrounding noise level.

For use before show time and for fillin in case of rain or accident, a record turntable and its preamplifier was incorporated in the console control cabinet. This unit could be plugged into the patch panel at the output terminal of the console and automatically replace the console program with the recorded program.

In addition to the stage sound system, a second independent amplifier and set of four 24" re-entrant drive unit loudspeakers were installed atop the grandstand roof to cover the outside plaza in front of the grandstand entrances. This system could be fed from the console, the turntable, the Celesta-Chime or from a microphone position located in one of the grandstand entrances. The primary purpose of this system was to advertise the pageant and also to furnish music to the plaza between shows.

The Deagan 25 note Celesta-Chime was chosen for this installation because it was found to be the only electronically amplified chime that could be played in tune with an orchestra. The chime rods are tuned harmonically which eliminates the "wow" and "out of tune" effect that is usually associated with chimes and bells. In some instances where the Celesta-Chime was played solo, full chords were used. No other instrument was found that was suitable for such usage. A special low "G" chime was built by Deagan to furnish the deep tolling bell effect that was desired for the Lincoln Funeral Train scene. As explained above, these chimes were reproduced through the control console and entire system and could be controlled the same as any microphone.

Since all the sound and control equipment had to be integrated with the entire operation of the show, the design and installation of the intercommunication, buzzer, cue and railroad signal systems was undertaken along with the sound system. The intercommunication system was in two stages; the first stage consisted of an electronically amplified 20 station master unit located in the director's booth connecting with remote stations back stage, ticket booths in the plaza, the orchestra pit, the equipment tents north and south of the stage, the electricians' switchboard room and the sound man. The importance of this intercom system can be realized when it is understood that it was the only two-way communication that the director had with the show from his remote position. It was used to check on equipment, cue in some of the vehicles, instruct the orchestra conductor and deliver the light cues, of which there were 62, to the electricians under the stands. As a supplementary system, there was a battery operated telephone type intercom system with three master stations. One was in the director's booth, one at the sound control console (visible at the extreme end of the control console, Fig. 3), and one in the orchestra pit. This system was used mainly between the orchestra conductor and the sound man.

The buzzers, music cue lights and the railroad signals were all controlled from a small custom-built consolette located in the director's booth. Buzzers were located in all dressing rooms backstage. in the wings and at the extreme ends of the stage extension wings. Each buzzer could be controlled independently or all could be sounded at the same time. The music cue lights were located on the orchestra conductor's stand and con-

PHOTO CREDITS

6, 7, 8.. Federal Telecommunication Laboratories

10...Official Department of Defense

sisted of one flashing white warning light and a second steady amber "go" light. The railroad signals were located at the ends of the stage wings and were used to cue the locomotives on stage. These signals had to be visible to engineers on any of the four switch tracks in either the north or south yards both in daylight and dark. The operation of the controls for these buzzers, lights and signals had to be simple and foolproof as a mis-cue during the show could cause a serious accident. Some of the trains that crossed the stage traveled 30 to 40 miles per hour, so it is easy to see that an accident would be hard to prevent in case of a mis-cue.

During the rehearsals before the show opened, a temporary talk-back system was installed in the grandstand to enable the directors to speak to the people on stage and in the equipment and railroad yards. This system utilized the two re-entrant projectors that were on either end of the grandstand roof and additional projectors that were placed in the stands and aimed toward the stage.

Much of the success of the entire installation was due to the intelligent discussion of the entire staging problem by Helen Geraghty, the director of the pageant. By talking over the entire show, and not just the sound system, a clear picture of the entire operation was established before the designing was even undertaken. It must be emphasized here that too many times a sound system of this type will be installed as just a P.A. job and will fail to do the job because "new and unusual" conditions pop up that had not been thought out. One of the reasons that this installation was so successful was that, not only was it a good sound reinforcing system, but a live part of the show. The two operators, who alternated at handling the mixing, worked from a complete technical script that listed all the dialogue, the music cues and the mixing cues. In all they had 260 separate sound cues to handle in one hour and eight minutes.

A good sound installation is noticed and appreciated by those who have to listen to it. Many compliments were received this past summer from musicians, directors and members of the industry who attended the show, and many favorable comments were voiced by members of the audience who had never before paid any attention to the sound systems that they had heard. In every large outdoor amphitheater, stadium or fairgrounds attention should be paid to all the details that contribute to or detract from the coverage and quality of the sound system installation so that high-quality sound systems can become the rule instead of the exception.

~@~

QUARTER-WAVE MATCHING SECTION

This chart is used to obtain the surge impedance of a $\frac{1}{4} \lambda$ matching section used as an impedance transformer from one real impedance to another.



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Out of my laboratory has come an entirely new Television Training...cutting months off the time required in old methods. I give all the knowledge and experience you need in weeks instead of months. I start where your present radio experience ends. The same day you enroll with me, I rush the first of many big Television kits that I will send during your training. From the first hour you are experimenting and testing practical TV circuits...and you keep right on from one fascinating experiment to another. You build the remarkable new Television Receiver-Tester illustrated at the left and useful TV Test Equipment. I give you theory, too, but it's 100% practical stuff that will make money for you in Television.



Exclusive THREE-UNIT Construction

Exclusive 1HREE-UNIT Construction You build my Television Receiver-Tester in three separate units-one unit at a time...each complete and self contained within itself. With each unit you perform dozens of im-portant experiments-and each unit may be used in actual Television receiver servicing. In this way my training may save you many dollars by eliminaring the need for costly TV Test Equipment. With these three units you can locate most TV Receiver troubles quickly and easily.

TV Tuner - I.F. Unit

Contains the RF amplified local Contains the RF amplified local oscillator, mixer and three stages of broad band IF amplification and the video second dector. The output constitutes the video signal and audio IF signal. For training, it is used to build and test video second detector, and stagger tuned IF amplifier obtaining 4.5 mc band pass. For TV servicing, it becomes a TV calibrator for IF alignment, substitute tuner, IF signal injector and second detector.

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If you are a radio-serviceman, experimenter, amateur or advanced student.. YOUR FUTURE IS IN TELEVISION. Depending upon where you live, Television is either in your town now... or will be there shortly. This is a vast new industry that needs qualified trained men by the thousand to install and service TV sets. There's really big money in Television, but you MUST know what you are doing to "cash-in" on it. I will train you in a few short weeks if you have had previous radio training or experience. had previous radio training or experience.



Video-Audio Amplifier Unit

Provides 4.5 mc IF ratio detector, low voltage power supply. For TV low voltage power supply. For TV, it becomes the audio output, includ-ing speaker, video output and low voltage power supply for RF and IF stages. For training, it is used to build and test transformer type power supplies, audio, video, IF amplification and FM detection. For TV servicing, it is an audio signal TV servicing, it is an audio signal tracer, IF signal tracer, video signal tracer and low voltage power supply.



Video Tube "Scope" Unit

scope unit contains low and high voltage (6000 V.) power supply for independent operation. For tele-vision, it becomes the sync, vertical and horizontal sweep circuits and their power supplies. For training, it is used to build and test most TV power supply, deflection, sweep, oscillator, and sync circuits. For TV servicing, it is a video signal tracer and sweep signal analyzer as well as substitute high and low voltage power supplies.

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statement, for their sales have proved the pull of the

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~	use Liemenis		B13-137X	1
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	B11-115	1	B11-139	2
ł.	B11-116	1	B13-139	1
	B17-116	1	B13-130Y	1
	B11-119	1	B13-137X	
	B11-120	1	Inner Shaft Ends	
	B11-121	1	E-187	3
	B11-123	2	E-190	1
	B 11-128	2	E-202	2
	B11-130	1	Sleeve Bushings	
	B12 120V	2	5-4	1
	B10 120X	,	5-5	- îl
	B18-130A		0.5	
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PROGRAM OF EVENTS

4	RADIO & TELEVISION NEWS
50 Parts Distribut	ors Conference & Show ends.
	Exhibition Hall Closes
6:00 p.m.	Registration Desk Closes
2:00 p.m.	agement"North Baliroom, Stevens Hotel
1:00 p.m.	Display Koom Area Opens
1 20	and all distributors
	Exhibition Hall OpensAttendance confined to ex- hibitors, their booth attendants, sales representatives,
10:00 a.m.	Registration Desk Opens—Exhibition Hall Lobby
URSDAY, May 25	
5.00 p.m.	Exhibition Hall Closes
6:00 p.m	Registration Desk Closes
2:00 p.m.	Distributors' Educational Session on "Inventory Con-
1:00 p.m.	Display Room Area Opens
	Exhibition Hall Opens—Attendance confined to ex- hibitors, their booth attendants, sales representatives and all distributors
10:00 a.m.	Registration Desk Opens—Exhibition Hall Lobby
DNESDAY, May 2	4
	Exhibition Hall Closes
6:00 p.m.	Registration Desk Closes
2:00 p.m.	Distributors' Educational Session on "Sales and Mer- chandising"—North Ballroom, Stevens Hotel
1:00 p.m.	Display Room Area Opens Distributors' Educational Service on "Sales and the
	and all distributors
	Exhibition Hall OpensAttendance confined to ex- hibitors, their booth attendants, sales representatives,
10:00 a.m.	Registration Desk Opens—Exhibition Hall Lobby
ESDAY, May 23	
7:00 p.m.	INDUSTRY KEYNOTE DINNER—Grand Ballroom, Stevens Hotel
6:00 p.m.	Registration Desk Closes
3:30 p.m.	Annual Meeting of Member-Exhibitors and General Members of Radio Parts & Electronic Equipment Shows, Inc.—South Ballroom, Stevens Hotel
2:30 p.m.	Parts Manufacturers, Inc.—Upper Tower Room, Stevens Hotel
2:20 a m	ufacturers; and West Coast Electronic Manufacturers' Association—South Ballroom, Stevens Hotel
2:00 p.m.	Lower Lower Koom, Stevens Hotel Joint Meeting of Sales Managers Club, Eastern Divi- sion: Association of Electronic Parts & Faultment Man-
12:30 p.m.	Luncheon Meeting for General Members—"The Rep- resentatives" of Radio Parts Manufacturers, Inc.—
10:30 a.m.	General Meeting—"The Representatives" of Radio Parts Manufacturers, Inc.—Upper Tower Room, Stevens Hotel
10:00 a.m.	Registration Desk Opens—Exhibition Hall Lobby
9:00 a.m.	Erection of Displays
ONDAY, May 22	"ASSOCIATIONS' MEETING DAY"
6:00 p.m.	Radio's Old Timers—Cocktail Party for Members— South Ballroom, Stevens Hotel
10:00 a.m.	Registration Desk Opens—Exhibition Hall Lobby
9:00 a.m.	Exhibition Hall and Display Room Area open for erection of displays
NDAY, MAY 21	



Use Wave Traps Made With OMMITE "Frequency-Rated" CHOKES

There are many cases where the front end of a TV receiver lacks the selectivity required to eliminate very strong 14 or 28 Mc signals. These fundamental frequencies jam through the receiver front end and create many kinds of cross-modulation products which affect the viewed picture (not to mention the setowner's temper). The only solution: each receiver affected must be treated separately, to prevent such fundamental frequencies from getting in and affecting the TV signal.

Tuned wave traps—either series, parallel, or combination types—are most commonly used to eliminate this trouble. They can be made effectively with Ohmite high-Q Chokes, in combination with suitable trimmer capacitors. (See diagrams at the right.)

Since most receivers have 300-ohm input, that type has been treated here. These designs, however, can be adapted to 72-ohm input as well. Scries-tuned traps are connected from the receiver terminals to ground. In some cases, however, two parallel-tuned traps may prove equally satisfactory. The chokes and trimmers can be mounted on a small terminal board with binding posts or clips for easy installation.

> WRITE FOR BULLETIN 133 AND CATALOGUE 21

OHMITE MANUFACTURING COMPANY 4885 W. Flournoy St., Chicago 44, Ill.





at

Table below shows approximate capacity necessary to resonate OHMITE Frequency-Rated Chokes

varioue	frequencies:	
various	nequencies.	

OHMITE CHOKE	mmf at 3.5 Mc	mmf at 7 Mc	mmf at 14 Mc	mmf at 21 Mc	mmf at 28 Mc	mmf at 50 Mc	mmf at 56 Mc
Z-28	98	25	6.2				
Z-50		74	19	8.2	4.6		10
Z-144			72	32	18	5.7	4.5
Z-235				69	39	12	9.7
Z-460	1					51	40



May, 1950

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These service managers know that modern FM and TV development and servicing demands test equipment made to the most exacting standards. They prefer the Simpson Model 480 Genescope because it is the most accurate, flexible, and convenient instrument available.

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In addition to providing all necessary signal sources, the new Simpson Genescope includes a high sensitivity oscilloscope of unique advanced design, complete in every detail. Equipped with a high frequency crystal probe for signal tracing. AM and FM oscillator sections provided with large, easy to read dials with 20:1 vernier control and 1000 division logging scales. *Revolutionary, ingenious, exclusive* output termination provides for various receiver impedances, either direct or through an isolating condenser. Step attenuator for control of output. Size $22^n \times 14^n \times 7\frac{1}{2}^n$. Weight 45 lbs. Shipping Weight 54 lbs.

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Band A: 2-120 megacycles Band B: 140-260 megacycles

Sweep width variable from zero to 15 megacycles

Sweep rate 60 cycles per second

Specially designed frequency sweep motor Continuously variable

attenuator

Crystal calibrator: $5 \text{ megacycles} \pm .05\%$

Audio Oscillator 400 cycles

Output Impedance 75 ohms

Step attenuator for control of output

OSCILLATOR Band A: 3.3-15.6 megacycles Band B: 15-75 megacycles Band C: 75-250 megacycles 30% modulation at 400 cycles or unmodulated Continuously variable attenuator Visual method of beat frequency indication OSCILLOSCOPE Vertical sensitivity: 35 my per inch Horizontal sensitivity: 70 my per inch Linear sweep frequency: 3 cycles to 60 kilocycles 60 cycle sine sweep Frequency essentially flat to 200 KC. usable to over 3 megacycles



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A trial of these better molded tubulars will convince you. See

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WATER TEST Far surpasses any existing specification reavirement.



HEAT TEST Excellent operation under high temperature conditions.

PULL TEST Leads resist breaking or pulling out. Takes rough handling.

AN ALMAN ALS LIFE TEST Long life even under most severe operating conditions.

SANGAMO ELECTRIC COMPANY SPRINGFIELD, ILLINOIS In Canada: Sangamo Electric Company Limited, Leaside, Ont.

May, 1950

27

Centralab Components



placement capacitors in repairing radio sets, television receivers and other electronic equipment, you can be sure of customer satisfaction. You can be sure of new business. What's more, Centralab capacitors — *Hi-Vo-Kaps*, Kolordisks, TC and BC Hi-Kaps — make your job easier, save you time. Here's why. They're simple to install . . . easy to stock. Get all the facts on profit-making CRL replacement capacitors from your nearby Centralab distributor.

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CERAMIC TRIMMERS for padder application in RF and HF circuits. These trimmers are noted for their great mechanical strength and electrical stability; low power factor. Hold circuit drift to a minimum. Truly an indispensable "must." Ideal for amateur, experimental and industrial use.



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MODEL "M" for voltage-divider, antenna shunt and "C" bias control, tone control, AF grid control. MODEL "1" for all miniature applications; rated at 1/10 watt, actually smaller than a dime. MODEL "R", wire wound, for voltage divider, antenna shunt, "C" bias, AF grid or tone control circuits.

SWITCHES

ROTARY for band change, meter, intercom circuits; made in ceramic and phenolic models. ROTARY SPRING RETURN for meter selection, intercom, phono-radio applications. MEDIUM DUTY for band changing in low power exciter-transmitters and receivers. LEVER ACTION for intercom, speaker, microphone and other applications.



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Outer tube plastic impregnated to prevent moisture-absorption

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Type 85TOC "Humidi-Seal" capacitors are specially designed for 85° C, operation, even in the most humid at mospheres, and will meet the severe present-day demands of endurance in television receivers, auto radios, etc.

WRITE FOR COMPLETE LITERATURE

Representatives and Distributors throughout the U.S.A. and Canada







JACK HARRISON has been named to the post of general manager in charge

of the custom-built cabinet division of *Mars Television*, *Inc.* of Corona, Long Island.

Mr. Harrison comes to his new post from W & d J*Sloane* where he served as manager



of the radio and television department for the Fifth Avenue department store. Prior to his association with *Sloanes* he was, for many years, the sales engineer for *Scott Radio*.

His new duties will include the creation of individually styled cabinets in addition to handling the sales and distribution of the company's custom line.

SIMPSON ELECTRIC COMPANY has recently merged with *The American Gage & Machine Company* and will be known as the *Simpson Electric Division.*

Ray Simpson, veteran of 38 years in the industry, will remain as chairman of the new division. Herbert Bernreuter has been elected to the post of vice-president of the parent company and will act as operating head of the *Simpson Division*, with which he has been associated since its inception.

Personnel, operating and sales pollicies will remain unchanged.

HAROLD J. NEWTON, vice-president and general sales manager of *National*

Electric Products Corp. of Pittsburgh, resigned his post recently because of ill health.

The resignation of Mr. Newton brings to a close a 38-year continuous sales effort for the

company. He was first associated with the organization in 1912 and transferred to Pittsburgh in 1929 as eastern district sales manager. He later served as western sales manager and sales manager of the corporation.

He has been vice-president and a director of the firm since June, 1948. * * *

AUSTIN LESCARBOURA, a well-known figure in the radio industry, has been awarded the French Legion of Honor for technical assistance rendered over many years and more recently to industrial missions sent to the United States in conjunction with the rehabilitation of postwar France.

This is the third French decoration

awarded to Mr. Lescarboura. He previously received the Officier d'Academie decoration in 1919, followed by his promotion to Officier de l'Instruction Publique in 1947.

CONCORD RADIO CORPORATION and *Radio Wire Television, Inc.*, have announced a severance of the joint ownership of stock in the two firms.

Samuel J. Novick has acquired all the stock and is now the sole stockholder of *Concord Radio Corporation* which operates stores in Chicago and Atlanta, Georgia.

Abraham Pletman has acquired all the stock and is now the sole stockholder of *Radio Wire Television, Inc.*, which operates stores in New York City, Newark, New Jersey, and Boston, Mass. Heretofore, the stock in each corporation was owned jointly by the two men. The stores will continue to operate under the same names as before.

CHARLES L. DAVIS has been named

manager of distributor sales to head the new distribu-

tor sales setup at American Transformer Company of Newark, New Jersey.

Mr. Davis was formerly associated with Jones Motorola Company of



Stamford as vice-president in charge of sales, service, and advertising. Prior to that association he was assistant to the vice-president of *Central National Bank* of New Rochelle, New York and assistant to the treasurer of the *Delaware, Lackawanna and Western Coal Company* of New York City.

Under the new program headed by Mr. Davis, distributors will be appointed in key locations throughout the country to handle the sale of *AmerTran* transformers.

* * *

INSTRUMENT ASSOCIATES has moved its office and warehouse to 37 East Bayview Avenue, Great Neck, New York ... HERMAN H. SMITH, INC., manufacturer of radio hardware, electrical components, and TV accessories, has moved to new and larger quarters at 436 18th Street, Brooklyn 15 . . . POL-ARAD ELECTRONICS CORPORATION has increased the size of its manufacturing facilities by moving to 100 Metropolitan Avenue in Brooklyn ... W. H. BRADY COMPANY recently moved to 204 West Washington Street in Milwaukee . . . The increase in television production has forced HAYDU BROTHERS, manufacturers of electron

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guns for television tubes, to acquire a plant in Mount Bethel, New Jersey, with a 10,000 square foot addition being erected at the present time . . SIGHTMASTER CORP. has moved to 111 Cedar Street in New Rochelle, N. Y. The new plant will enable the company to quadruple production of its television receivers . . . ACCURATE MANUFACTURING COMPANY, manufacturers of friction and rubber tapes. has recently completed a new plant in Garfield, New Jersey, which will in-crease the facilities of its existing factory building in that city . . . Construction has begun on a new 15,000 square foot area building in Cuba, New York, which will house additional production facilities for ACME ELEC-TRIC'S line of television transformers . CORNING GLASS WORKS is building a new plant in Albion, Michigan, for the manufacture of glass bulbs for television picture tubes . . . HOFFMAN RADIO CORP. has added 20,000 square feet of floor space to its plants in Los Angeles . . . GLOBE-UNION, INC., has acquired a new plant in Denville, New Jersey, which will be devoted to the production of a full line of ceramic capacitors manufactured by the Centralab Division of the company . . . The new television tube plant of the RADIO CORPORATION OF AMERICA was recently dedicated at Marion, Indiana, with local and state dignitaries in attendance.

GEORGE R. SOMMERS has been appointed assistant general sales mana-

ger of the Radio Tube Division of Sylvania Electric Products Inc.

Mr. Sommers joined the company as a lighting fixture specialist in 1940 and early in 1943 was transferred to



the Washington, D.C. territory where he instituted a decentralized sales policy which was later adopted for the company's district sales offices. He was appointed director of Pacific Coast sales for *Sylvania* in 1945 and in 1949 was transferred to the New York office to assist the general sales manager of the Radio Tube Division.

WALTER F. GREENWOOD has been named manager of industrial sales for Telechron, Inc., of Ashland, Mass. . . . PINCKNEY B. REED is the new manager of the Industrial Equipment Section of the RCA Engineering Products Department . . . Associated Merchandising Corporation has named GEORGE W. SMITH to the post of merchandise representative for radios, television, and records . . . D. Y. ROBINSON will represent Webster Electric Company's sound division in the New York area . . R. C. DOVE and H. V. SOMERVILLE have been named operating assistant to the vice-president and technical assistant to the vice-president respectively with the distribution division (Continued on page 120)

RADIO & TELEVISION NEWS

For Clear Sharp Pictures and Powerful Trouble-Free Performance



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Will Handle 7/8" to 2" diameter Upper and Lower Masts

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- Heavy-Duty Motor Reverses Instantly
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Available in the Following Models: • TR-1...rotator with control box \$37.45 • TR-2...compass control rotator with • TR-2...\$44.95 "Perfect Pattern" dial\$44.95

A Combination You Can't Beat ... It's a C-D Sky Hawk Antenna with a C-D Sky Hawk Tele-Rotor

May, 1950

The television industry has been looking for something like this ... and it took CORNELL-DUBILIER to produce it! THIS is the last word in ROTATORS ... the finest! ... BECAUSE IT HAS EVERYTHING! It wasn't developed overnight but is the result of fourteen solid months of research and development! BUT ... it has been worth it! ... BECAUSE IT MEASURES UP TO CORNELL-DUBILIER STANDARDS



C-D SKY HAWK

TELEROTOR

CONTROL

... with the PERF PATTERN dial... finge control with instant I cation of antenna posi shown on the illumine two-tone face.

* hallicrafters

1950 Version of the famous World War II (and pre-war) Transmitter!

The communications "workhorse" of World War II has returned to its peacetime role! Limited quantities of this most respected member of Hallicrafters transmitter family are now being built for the civilian market. All transformers hermetically sealed. Shock mounted for mobile use. Complete with tuning units and output coil sets for 2 to 18 Mc (coils for 28-30 Mc available separately). 450 Watts CW, 300 AM phone. Ask wherever Hallicrafters communications equipment is sold or write to the factory for a spec sheet.

HT-4G..\$1,520.00

Speech Amplifier HT-5G....\$217,00



3.32 -10

"The Radio Man's Radio"

See and Examine the Latest Hallicrafters Equipment at the Radio Parts Show—Chicago—May 22nd to 25th.



By J. R. POPKIN-CLURMAN, W2LNP Hazeltine Electronics Corporation

> Fig. 1. To simplify construction and testing, system is built in separate units.

ROM time to time there have been amateur television systems - described in the various radio publications. While these have been of good design in most cases, all lacked the definition and quality that people have been led to expect since the advent of popular television. The system to be described is capable of the full resolution and detail associated with the present day television standards.

By the ingenious use of scanning pulses taken from any standard television broadcast station by means of

a standard TV receiver, much of the complication of a commercial transmitter is avoided.

The expensive image orthicons and iconoscopes are replaced by standard cathode-ray tubes which may be of the surplus variety.

Reception of the signal is accomplished by any TV receiver equipped with a simple converter. If desired it is possible to use the same receiver to pick up both the scanning pulses

from the transmitter, and for reception of the transmitter picture from another amateur TV station.

Operation is on the 420-450 mc. band, permitting the use of small high gain antennas and readily available tubes.

While the equipment may appear complicated, it has been broken into several component parts to simplify the construction. This enables the advanced amateur to construct and test Part 1. Complete description of a ham TV station that can be built from available parts at relatively low cost. Succeeding articles will present details on constructing entire station.

the units individually and avoid many of the pitfalls associated with a combined unit.

It should be pointed out that the equipment is operating on an amateur band and only licensed amateurs may

will have many of the parts needed and be able to hold the completed cost to a low figure. In any event the cost will be far below that of any comparable system.

This television system uses relatively common com-

ponents and is capa-

ble of being built by

the average experi-

menter experienced

in the radio art. It

provides a means of

making a standard-

definition broadcast picture of 525 lines,

60 fields, 30 frames interlaced two-to-one,

having a video band-

width up to 4.5 meg-

acycles. In this sys-

tem, for simplicity, a

single transmitter

sends the sound si-

multaneously with

the video by means

of frequency modula-

BOTH SOUND AND VIDEO ARE INCORPORATED IN SINGLE TRANSMITTER STANDARD DEFINITION, 525 LINES, 60 FIELDS, 4.5 MC. BANDWIDTH IS PROVIDED NO EXPENSIVE ICONOSCOPES OR IMAGE ORTHICONS REQUIRED STATION COVERED OPERATES ON 420 MC. BAND. HIGHER FREQUENCY AMATEUR BANDS CAN BE USED SYSTEM DESCRIBED IS FOR TRANSMISSION OF STILLS. ALTERNATE METHOD FOR SCANNING MOVING OBJECTS IS DISCUSSED IMPROVED SYSTEM, ALSO DESCRIBED, EMPLOYS STANDARD TV RECEIVER TO PROVIDE SYNC AND SOUND FROM ANOTHER AMATEUR STATION RECEIVE PICTURE AND SOUND FROM ANOTHER AMATEUR STATION TO REDUCE COST ANY STABLE 420 MC. TRANSMITTER, EITHER

TO REDUCE COST, ANY STABLE 420 MC. TRANSMITTER, EITHER SURPLUS OR HOME-BUILT CAN BE USED

OPERATOR OF EQUIPMENT MUST BE A LICENSED AMATEUR. TO BUILD COMPLETE STATION CONSTRUCTOR SHOULD BE FAMILIAR WITH RADIO DESIGN AND THEORY

operate this television equipment. This first article will outline the system used and show different ways of accomplishing the final result. Succeeding articles in this series will give details of the various component parts which go to make up the complete station.

Many of the components used are taken from surplus equipment to reduce the over-all cost. Most amateurs tion of a 4.5 megacycle sub-carrier. The complete signal is transmitted on an amateur carrier frequency, such as 432 megacycles. Reception can, therefore, be accomplished on a standard television receiver preceded by a simple frequency converter.

The basic "picture maker" or camera used in this inexpensive system is derived from an old device used in early television experiments with ro-



Fig. 2. Block diagram of transmission system showing functions of the units.



Fig. 3. Method for using cathode-ray tube as light source for 35mm transparencies.



Fig. 4. A 420 mc. converter is used with α standard television set for reception.

tating discs and known as the *flying* spot scanner.

Suppose that a spot of light such as is generated by an arc lamp is fed through a rotating disc with small holes in it arranged spirally so that they progressively become nearer to the center of the disc. This means that if a scene is illuminated by this light,

Fig. 5. Diagram showing direct reception of scanning signals from a TV station.



there will be one small flying spot after another traversing the scene, each illuminating a line of the scene, and each line successively displaced, so that at the end of one revolution of the disc the entire scene has been covered. As the spot of light covers each small portion of the scene, a phototube picks up the light derived from this particular portion of the scene. The phototube current is then a video signal suitable for transmission. (At the receiver the corresponding signal is amplified and drives a source of light which can be modulated in intensity. This light is then viewed through a similar receiving disc rotating in synchronism with the transmitting disc. allowing the original scene to be observed.) This old principle is applied electronically in the modern flyingspot scanner. Instead of using the rotating disc with holes, a raster is generated on the screen of the cathoderay tube shown as unit (2) in Fig. 2. Since the raster is produced by the motion of one dot behind the transparency, the variations of light value from point to point of the transparency are picked up by the phototube.

In the present system the transmitter uses an ordinary TV receiver altered so as to separate the sync from the received television signal from any local TV station. This sync is then used to generate the interlaced standard RMA sweeps. A block diagram of the sending end of the system is given in Fig. 2. Unit (1) is the broadcast TV set, which is standard and whose

sweeps and sync-separating circuits are as normally found. It is tuned to any local TV broadcasting station. In the event that it is not possible to receive a standard TV signal, then it is merely necessary to derive the sync and blanking from the TV set's own sweep circuits. In this case the picture will often have only 262 lines noninterlaced, since the horizontal oscillator is running free. The picture retains the same horizontal resolution but the vertical resolution is reduced by a possible factor of two. For this type of operation it is desirable to sync the vertical to the 60-cycle power supply so that hum effects will be reduced.

The lead to the video amplifier from the grid of the television set's cathode-ray tube (2) is disconnected in the set and is instead connected to the blanking generator (8). (For this source (2) of the flying-spot illumination, practically any CRT can be used, including surplus P7-phosphor radar types.) In addition, the sync derived from (1) is to be used later for transmission purposes in mixer (7). The light from the raster of (2) is then passed through a transparency (3) (which is the picture being transmitted). This can be held in a slide projector holder or put directly on the face of the cathode-ray tube (2). If a slide projector is used the phototube is substituted for the slide projector light source, as shown in Fig. 3. Returning to Fig. 2, the light passing through the transparency is picked up by a phototube (4), amplified by a video amplifier (5), passed through a video clipper and blanking inserter (6), and then passed through a sync and sound sub-carrier mixer (7). The output of the mixer is fed to a combined videosound modulator (10), which modulates a transmitter (11), and this feeds an antenna radiating the picture and sound signals simultaneously.

The receiver block diagram is given in Fig. 4. The combined picture and sound carrier is picked up by a converter (12), which includes a "cascode" (grounded-cathode triode followed by a grounded-grid triode) i.f. amplifier as its output section; it delivers an output corresponding to an unused channel of a regular standard broadcast TV set (13), where the picture and sound are separated by the normal processes.

Other possibilities lend themselves suitably to this system. In Fig. 5 the TV station A supplies picture and sync for the amateur television transmitting station located at B. If the receiving station is located along a circumference of radius AB, say at C, it becomes possible for the receiver Cto use the sync directly derived from station A without having to receive sync from transmitter B. This means that all that is necessary to receive B's transmission is to disconnect the video from the picture-tube grid of C's receiver and use instead the video received from B over the u.h.f. link. The broadcast TV receiver at C is tuned

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to station *A* and is, therefore, scanned in synchronism.

Under these conditions reception of B's picture at C will be accompanied by a horizontal shift or displacement of the picture proportional to the delay with which B's picture arrives at C with reference to the sync received from A. For short distances this shift is small. For greater shifts recentering of the picture at C or adjustment of the horizontal sync phasing control becomes necessary.

This system is not limited to transparencies. It is possible to send pictures of live subjects or of fixed objects as shown in Fig. 7. The raster from the CRT is projected by a projection lens on the object, and the reflected light from the object is collected by a condensing-lens system for delivery to the phototube. The remaining operations are then just the same as with a transparency. Considerably greater light is needed than for the transparencies, and a 5TP4 or 5WP15 projection-type CRT with associated high voltage is suitable.

For sending motion pictures the setup is essentially the same as for still transparencies. A film projector has its light source removed and the phototube put in its place. A 60-cycle synchronous motor is used to drive the film sprocket and the film is run through the projector at 30 frames per second instead of 24. This speeds up the motion on the film somewhat. It is also necessary to blank the raster during the film pulldown time. More elaborate means are needed to show the film at 24 frames per second.

Since the transmitter (unit (11) in Fig. 2) sends out the picture doublesideband, a total bandwidth of nine megacycles at the transmitter is desirable. This presents no problem on 420 megacycles. However, in order to use a standard receiver advantageously it is preferable to follow the vestigial-sideband method as in broadcast practice. One TV sideband with the vestigial other sideband is obtained by the selectivity of the receiver's i.f. amplifier, resulting in reception and reproduction of standard RMA pictures. See Fig. 8. The shaded portion represents the sideband eliminated by the selectivity of the receiver. Of course in tuning up the transmitter it is also possible to adjust the tuning in such a way that one sideband is amplified more than the other-naturally the sideband to be used in the receiver should be favored.

Since the sound is transmitted as a sub-carrier, it becomes impossible to use full modulation on the transmitter for pictures as this would not allow the sub-carrier of 4.5 megacycles to be continuously transmitted. Generally fifty per-cent modulation is used for the sideband carrying the FM subcarrier and the other fifty per-cent devoted to video signal. Changing these proportions of modulation in favor of more picture modulation will result in undesirable buzz in the sound portion of the transmission. Of course, if it is



Fig. 6. The original transparency (left) compared to the reproduced image (right).

desired, a separate sound transmitter spaced 4.5 mc. can be used, or other sound not necessarily associated with a picture can be used.

The phototube used is a standard surplus 931A multiplier type. These were used in most of the radar countermeasure noise generators for radar jamming (AN/APT-5). In the noisegenerator application, use is made of the extreme amplification capabilities of the tube. However, it is not desirable, in the present case, to employ enough gain to reach the noise-level capabilities of the tube, as this results in "snow" in the television picture. For convenience in handling the total of 550 volts supply potential, a positive and negative source of voltage is used. The gain control of the whole video system is obtained by adjusting the value of the negative supply voltage.

The output of the phototube is fed to a series of video amplifiers, including one stage, known as the high-frequency peaker. This peaker is necessary to equalize or compensate for the time it takes a spot of light to build up and decay, as the phosphor of the cathode-ray tube (2) does not light up or die down instantly. This corresponds to a pulse of voltage in the phototube which is exponential in rise and decay time. This is the same as saying that the pulse derived from the spot of light is deficient in high frequencies, which are then equalized by the additional high-frequency response of this stage. The desired rise and fall time of the pulse is shown in Fig. 9 by the dotted line. The actual pulse without the high-frequency peaker is shown by the heavy lines. If the peaker is used, the higher video frequencies are favored so that the dotted wave shape is obtained. This effect is produced by an RC network. As the condenser C is adjusted, the "blurs" and smears in the video picture will disappear causing the picture to sharpen up and display the detail expected.



Fig. 7. Method of using cathode-ray tube as a light source for scanning live subjects.



Fig. 8. The carrier as transmitted and the vestigial sideband obtained (right).

It is necessary to provide for picture reversal in order to transmit from either a negative or positive picture transparency. The received picture will then always be positive.

The blanking portion in the RMA signal cannot generally be used be-

Fig. 9. Compensation in the video stage.





Fig. 10. The regulated supply used to power all units except the r.f. section.

cause the black portions of broadcast pictures are sometimes so close to the blacker-than-black region of the composite signal that the flying-spot raster might have some unwanted black portions. However it is possible to use the horizontal sync pulses as H blanking pulses. A better way is to use the flyback period of the H and V oscillators in the receiver for deriving blanking signals. It is usually possible to find some place in the sweep circuits which already has a good pulse of the proper polarity. For example, the vertical pulse present at the output of the vertical integrating network. The horizontal blanking is easily obtained from the output of the sync separator or by

taking a portion of the flyback pulse present in the horizontal deflection coils. These signals are then fed to the blanking generator where they are clipped and then fed to the flying-spot CRT grid and the video blanking inserter.

So far in our discussion, the video has been allowed to go through the video amplifier without thought of what should be done with the blanking and the sync. Actually the video is limited by superimposing on it a pulse which will shut off the picture-tube grid when the time comes for the retraces of the lines to take place. This is accomplished by adding together in unit (6) of Fig. 2 the video signal am-

Fig. 11. Photograph of the power supply of Fig. 10. Three outputs are available.



plified by (5) and a pulse derived from the blanking generator (8). The extremity of the blanking pulse with its video superimposed is then chopped off in the same unit.

Beyond the clipped portion, the sync signals derived from the TV broadcast (or locally generated) are superimposed in mixer (7). The result is a standard RMA experimental signal capable of being received on a standard receiver without any difficulty.

The FM sub-carrier unit (9) consists of a 4.5-megacycle oscillator modulated by a reactance tube. A portion of the 4.5-megacycle output is mixed with the video and handled then in the same manner as if it were a video component.

The video modulator is a straightforward amplifier similar to the power video stage of an ordinary television set. However, it has to swing over a much wider voltage range than is required in a normal television set. The output of a 6AQ5 feeds a 6AS7G power-handling cathode follower which is used directly to modulate the transmitter. The use of the cathodefollower modulator, with its low source impedance, is especially desirable because of the high bypass and stray capacitances generally associated with the transmitter.

It is desirable to use a regulated power supply for most of the circuits, associated with the video, scanning, blanking, and related circuits. For those anxious to get started, a satisfactory supply is illustrated in Fig. 11 with the schematic diagram shown in Fig. 10. Many amateurs will have a supply available which will meet these requirements.

Succeeding articles will deal with the construction of the various units and cover the tuning and adjustment. The experienced amateur will be able to make many changes in the basic design to accommodate equipment on hand.

For those who desire more power, a transmitter of the SCR-522 type will serve as a driver for a tripler furnishing an output of ten to twelve watts on the 420 mc. band.

There are several pieces of surplus equipment which operate on the 420 mc. band and if some of this equipment is on hand, many of the components may be used. Usually the receiver portion of this surplus equipment is of the superhet type with a broad i.f. system, and by substituting a front end capable of tuning to a TV station, the necessary signal for scanning may be obtained. In many cases there will be unused tubes which may be adapted to serve as the vertical and horizontal oscillators, or the video section included in the receiver may be used.

Of course operation is not limited to the 420 mc. band and the system may be used on any of the amateur bands above this frequency. The 420 mc. band was chosen as it permits the use of familiar techniques without resorting to waveguides and plumbing.

(To be continued)

RADIO & TELEVISION NEWS

Something NEW in TRANSCEIVERS

By KENNETH M. KORTGE, W8AHT

Fig. 1. Battery and audio unit with assembled handset. The r.f. section and antenna are mounted on earphone end of handset. A slot was cut in the shield can to clear projecting antenna socket. Cover is held in place by sheet metal screws.

Details for building a handy two meter transceiver from a readily-arailable surplus TS-13 field phone.

171110

TITH conditions becoming crowded in the larger cities and with more and more surplus crystal-controlled transmitters being put in use, the modulated oscillator is rapidly becoming a thing of the past on the two meter band and rightly so. But for super-portable, short-range operation the "handietalkie" version of the standard transceiver is still holding its own.

The common method of constructing the handie-talkie is to assemble microphone, headphone, batteries, antenna and all the parts in a single unit. This necessitates handling a fairly bulky package and struggling to manipulate switches as well as supporting the weight of the batteries, which, though relatively light, can tire the operator's arm if used for any length of time. The author decided to make an attempt at eliminating these two evils. The end result is the apparatus shown in the photographs.

To facilitate handling, a surplus TS-13 field telephone handset was used as the basic unit. This unit has a 200 ohm microphone, and a high impedance headphone. Many similar units are available. In order to keep the antenna as high as possible during operation it was decided to mount the r.f. portion, along with the antenna, on the earphone end of the handset. To house this part an aluminum shield can about two inches square and three inches tall was salvaged from the junk box. A base was cut from one-eighth inch aluminum and shaped so that the shield can would fit snugly over it. A flat was filed on the earphone end of the handset in such a way as to lie vertically when the phone is in operating position. The base was then fastened to this flat with machine screws after drill-

ing and tapping two holes, 34" apart, in the filed portion. A third hole was drilled between the two tapped ones through the base and the bakelite into the chamber occupied by the headphone to accommodate the leads from the r.f. section. Bringing these leads into the switch opening in the handle presented a problem since a hole could not be drilled directly from one opening to the other. The solution was to drill from each chamber toward the next at as close an angle as possible so that the two holes met. A stiff wire was pushed through the resulting "V" shaped path and the small conductors from the r.f. section pulled through with this wire. The same procedure was used to pass several wires from the switch chamber into the microphone opening. (There were conductors moulded in the plastic of the handset but most of them were severed during drilling operations so it was necessary to install replacement wires.)

The switch was found to be a double-pole, single-throw one of the leaf variety and, since a ganged d.p.s.t. and s.p.d.t. is used in the circuit, a (Continued on page 106)





Frank Andrea-

RADIO and TELEVISION PIONEER

Thirty years of progress in the radio industry is typified by the life and work of this pioneer.



Bv

ARTHUR PINE

O HISTORY of the radio or television industry would be complete without the saga of Frank A. D. Andrea, for during the 60 years of his life, this pioneer has added much to the advancement of electronics.

Today, as president of the Andrea Radio Corporation of Long Island City, New York, he is still busy each day with experimental work directed toward perfecting television to the point where it is flawless.

1950 is a landmark in his career for it marks his 30th year in radio and television, and places him, along with Lee DeForest and others of that caliber, as one of the pioneers in the industry. Few men still active in radio

and television today were in this field back in 1920 when he organized *Frank A. D. Andrea, Inc.* in a small room on Jerome Avenue at 176th Street in the Bronx.

At that time, the entire organization consisted of Frank Andrea and two helpers. Their first efforts were concentrated on producing such items as rheostats, crystal detectors, and sockets for the pioneer set builders of that time.

His interest in radio was stimulated by two of the pioneers in the communications field—Guglielmo Marconi, the "Father of Wireless" and Lee DeForest, one of radio's early "masterminds."

In 1917 he went to work for Lee DeForest. Then in July of 1920, *Frank A. D. Andrea, Inc.*, came into existence. Broadcasting soon came into its own and the company started to expand. By 1922, the organization had begun to prosper, a large percentage of the business being in "Neutrodyne" kits which were sold to amateur set builders for home construction. These were the first radio kits to be offered to the amateur, according to Mr. Andrea.

In 1923 the company produced the first commercial "Neutrodyne" receiver, based on designs by Professor L. A. Hazeltine of Stevens Institute of Technology. By March of that year the new receivers were being turned out in quantity and shipments were on their way to jobbers throughout the country. This receiver captured the pub-





A typical Andrea "Neutrodyne" set of the 1925 era.

The 19" "Normandy" model, one of the latest Andrea television receivers, with AM-FM radio and an attachment for a record player.

Frank Andrea with one of his engineers, Louis Clement (now Director of Research and Engineering for Crosley Div. of AVCO Mfg. Co.) working on their first cone speaker in the year 1925.



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The present-day Andrea cabinet assembly line where the completed chassis are installed into the television cabinets.

lic's fancy and over 30,000 were put on the market before the year was out—an undreamed of output for that era.

By the end of 1925 the company was well on its way to success. Throughout this period and until 1932, Mr. Andrea was interested in the development of television and spent a great deal of time and money developing the scanning method of television. This system used a revolving disc to obtain a picture. In 1931, the National Broadcasting Company was telecasting experimentally with the picture being received on television sets via the scanning method.

In 1932, Mr. Andrea was bitten by the "retirement bug," and he relinquished control of the company. It took only two years, however, for him to realize that his interests were too closely allied with radio and electronics to retire, so in 1934 he organized the *Andrea Radio Corporation*.

After much experimentation the first *Andrea* television kit was offered to the public in 1938, and, like the original "Neutrodyne" kit, was sold to those people who wanted to "assemble their own." Over 5000 of these kits were sold in the first year they were offered.

Today Mr. Andrea and his staff of engineers are busily engaged in the problems involved in color television. Like others in the industry, Mr. Andrea believes that the ultimate in television will be color and a third-dimension picture. -30-



The television assembly line at the Andrea plant. Here table model video receivers are being produced in quantity.



The original Andrea "Knockdown Neutrodyne Receiver Kit."

(Left) Scene at the original Andrea factory in the Bronx shows engineers testing one of the first Neutrodyne sets in 1923. (Right) Frank A. D. Andrea demonstrating one of his early Neutrodyne chassis to a group of Canadian engineers.



May, 1950



NDOUBTEDLY the most accurate method of aligning television receivers is with an os-

vision receivers is with an oscilloscope, sweep signal generator, and marker generator. The procedure is, however, rather involved and timeconsuming. There is also the possibiliity of error caused by loading of the circuit with the signal generator. In addition, many service technicians find the cost of test instruments a decided handicap. Furthermore this type of alignment procedure is not adaptable to servicing in the field, in homes, or away from the test bench thus entailing the time and trouble in removing the receiver to the shop.

Over a period of years the author has been developing and using a method of television alignment which has proven to be practical, inexpensive, rapid, and accurate within the tolerable limits of visual capabilities. This system has the added advantage of permitting the simultaneous adjustment of the linearity, aspect ratio, focus, and bandpass. No further adjustment of the picture position, size, definition, aspect ratio, focus, or linearity is required after aligning by the method outlined. With the control of the picture present, the technician is assured of the greatest gain, best available frequency response, and the most pleasing picture possible.

A simple method for aligning video receivers without the use of tools or test instruments.

Television alignment without the use of instruments is entirely possible providing the service technician understands basic television circuits and uses his knowledge wisely.

Test Pattern

The test pattern, an integral part of every station's programming, is the basis of the alignment procedure outlined. The test pattern should not be treated as just another picture because it is, in reality, the "last word" in what a properly aligned television receiver should be able to reproduce. In fact, in the final analysis, a good alignment job is one that produces greatest clarity and the finest detail in the test pattern.

In Figs. 2 and 4 are shown typical resolution charts. The more nearly the receiver reproduces these charts





the more perfect is the receiver performance. In order to better understand these resolution charts, a brief review of their characteristics is in order. Notice that the chart of Fig. 2 consists of two circles. These circles are used to obtain the proper aspect ratio, i.e., height to width ratio. When the circles are perfectly formed and fill the tube face while remaining circular, the largest possible picture is obtained, compatible with the proper height and width. An improper setting here may cause distorted letters and figures. The linearity and aspect ratio must be set correctly before adjustments are made to obtain the proper bandpass.

Next notice that part of the chart which is in the form of a Maltese Cross. The cross is made up of a number of converging lines. Each line is wide at the periphery of the circle and narrows as it approaches the center circle. At the periphery the left-hand and lower wedges represent about 150 line resolution. This means that approximately 150 lines of that thickness, with equal spaces of the same thickness between lines, will fill the horizontal or vertical space completely and be of the definition visible on the screen of the set. Now the

lines become narrower as they approach the center of the test pattern and the spaces between the lines also become narrower, occupying the same space between lines as the line width at each point on the pattern. At the center circle the lines represent 350-450 lines, varying with the pattern. This indicates that the receiver is capable of reproducing 350-450 lines and spaces, vertically and horizontally. Naturally the more lines and spaces a receiver can distinguish, the finer the detail it will be capable of reproducing. Fine detail is the result of optimum amplification of the full 6 mc. television bandwidth and particularly that part of the bandwidth associated with the video (5.25 mc.).

Notice, too, that the pattern contains wedges of varying degrees of shading in addition to the Indian head. These are used in adjusting the contrast and brightness controls to provide a full range of contrasting black to grey to light grey to white. The proper setting of the brightness control almost always requires a slight compensation in the contrast control. If the four distinct shades of the pattern are not visible in the wedges it is an indication of nonlinear distortion in the video amplifier. This is usually caused by an improperly biased or gassy tube. The signal from second transmitter should be а checked to insure that this trouble is not in the transmitter.

Focusing of the test pattern is very important and since present-day cathode-ray tubes do not focus evenly over the entire face of the tube a median setting must be obtained so that the most satisfactory over-all focus is achieved. This can be set tentatively as the point at which the lines of the raster are distinguishable and are of the minimum possible width.

Alongside of the wedges (Fig. 2) are a series of numbers from 20 to 45 (indicating 200 to 450 lines). These numbers are indicative of the number of lines represented by that particular line and space thickness. It is of great advantage not only to know the number of lines a receiver is capable of producing (as indicated on the test pattern) but also the frequency represented by that number of lines in order to check the definition as well as the lowest and highest frequencies capable of being received clearly. Table 1 is a handy chart which lists the number of vertical lines and the equivalent frequency in megacycles. The chart may be used in judging the bandpass of the resultant adjustments to the receiver.

The diagonal lines for interlace check enable the user to test for improper interlace. A television picture is scanned with alternate lines and the lines must interlace perfectly. An irregular saw-tooth effect on the diagonal lines indicates poor interlace caused by an improperly functioning vertical sweep system.

The horizontal lines of varying

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Fig. 2. The familiar Indian head test pattern used by many television stations.

length at the bottom of the circle are for the purpose of checking low frequency response. They should be sharply defined and without variation in intensity. Any variation in the intensity of these lines will indicate poor low frequency response.

The group of vertical rectangular areas located on either side of the inner circle are used to check "ringing" caused by overcompensation at the high frequencies. If the video output stages are overcompensated, these rectangles are followed by multiple images similar to ghosts but spaced equally from each other.

Heterodyne

Few people realize the dramatic changes which take place in a superheterodyne receiver during the heterodyne process. Prior to the advent of television little attention was paid to the high or low frequencies since in broadcast work most of the channels are so narrow that they can be easily tuned and the complete band of frequencies is always transmitted. Although many persons have a fairly

VERTICAL	FREQ.
LINES	(mc.)
100	1.25
150	1.56
200	2.50
250	3.12
300	3.75
350	4.37
400	5.00
	-

Table. 1. Chart for determining the frequency from number of lines on pattern.

sensitive ear for alignment, often a broadcast set receives only a portion of the transmitted signal. Tuning in a broadcast signal emphasizes the lows or increases the hisses and other high frequencies on the sides of the carrier. This same principle can be used not only in the alignment of the video section but in the proper adjustment of the sound portion of the television receiver.

Whenever two signals are mixed together in a nonlinear device two new signals are generated, *i.e.*, the original two signals are still in evidence but the presence of two new





signals is always apparent. These signals are the sum and difference between the two original frequencies. Therefore, if a 100 kc. and a 10 kc. signal are mixed in an amplifier, the r.f. output stage will contain signals of 100 kc., 10 kc., 110 kc. and 90 kc. If an r.f. stage follows (as in the final stage of the video transmitter) the only signal present on the antenna will be 100 kc., 110 kc., and 90 kc. The original low frequency signal is attenuated due to the limited bandwidth of the r.f. stage. Fig. 1 demonstrates this principle.

Thus it may be said that a 10 kc. signal represents a full bandwidth of 20 kc. (from 90 kc. to 110 kc. in this example) and in order to produce this 10 kc. signal properly *all* of the bandwidth should be received and amplified. In television an attempt is made to approach this ideal. Fig. 3A shows the ideal television signal for Channel 3 (60 to 66 mc.).

However, for the present, our concern will be with the video. Since the highest signal produced will be in the region of 4 mc. it is reasonable to assume a bandwidth of 8 mc. But experiment has shown that a great.deal of space may be saved by transmitting only a portion of the difference frequencies (that portion below the carrier frequency is referred to as the difference frequency while that above the carrier is known as the sum frequency). The portion transmitted is only 1.25 mc. of the original 4 mc. difference frequency. This signal can, however, produce an intelligible video picture and yet maintain a minimum of bandwidth. This principle is known as vestigial sideband transmission. This is the reason that the video carrier appears as it does in Fig. 3A.

A careful study of Fig. 3A reveals that the center of the video carrier is 61.25 mc. Low frequencies are gathered close to this point while the high video frequencies are farther away (both higher and lower than 61.25 mc.). Referring again to the test chart and frequency table it may be seen that a few kilocycles on either side of 61.25 mc, corresponds to definition in the horizontal wedges whereas high frequency video or vertical wedge detail is located at points well over 1 mc. and as high as 4 mc. on either side of the 61.25 mc. carrier. However, for video frequencies greater than 1.25 mc, the picture information is located in the sum frequency region of the video. Now add an oscillator signal, i.e., mix with this r.f. carrier from the station a local oscillator of fixed frequency as is the case in the superheterodyne process.

In the superheterodyne process an oscillator of fixed frequency is used and mixed with the incoming r.f. signal thus taking the difference frequency as the signal we call i.f. or intermediate frequency. Fig. 3B represents the video signal with the oscillator.

Notice that for Channel 3 the oscillator is tuned to 87 mc. Thus the center of the audio band is 21.25 mc. below and the center of the video band is 25.75 mc. below the oscillator. This is the reason for the two i.f. frequencies, 21.25 mc. and 25.75 mc. This choice of i.f. frequencies varies from receiver to receiver but in all cases the video and audio i.f.'s are separated by 4.5 mc.

Up to this point the processes do not materially differ from broadcast methods. Reviewing the r.f. video spectrum again and comparing the i.f. as is diagrammed in Fig. 3, the low frequency or horizontal wedge information is located at 61.25 mc. and extends a few kc. on either side. The i.f. low frequency horizontal wedge information is located a few kc. on

Fig. 4. The television resolution chart adopted as a standard in 1946 by the RMA.



either side of 25.75. Note too that the frequencies up to 1.25 mc. are gathered about the center carriers as before. This frequency will produce about 100 lines of vertical wedge information. We are striving, however, toward 400 line vertical wedge information which lies in the r.f. frequency range of 62.5 mc. to 65.25 mc. and in the i.f. range of 21.75 mc. and 24.5 mc.

This is very important and may be stated in various ways. The high frequency r.f. video is, when heterodyned, transformed to low frequency i.f. video. Note that the high frequency r.f. is the low frequency i.f. and that the low frequency r.f. is high frequency i.f. and that the higher the desired definition, *i.e.*, the finer the vertical wedge definition required, the lower the i.f. frequency must be. In short, to tune for low frequency video, 100-200 lines, iron slugs are pulled out of the coil forms. For high frequency video the slugs are turned into the center of the coil to increase the inductance and lower the i.f. frequency.

Remember that as more and more vertical wedge is required, the inductance and capacitance of the i.f. amplifiers must be increased. Also in order to produce greater clarity in the horizontal wedge, inductance and capacitance must be decreased.

Presetting the I.F.'s

Presetting the i.f.'s is necessary only when the receiver is completely misaligned. Set all the slugs in the video i.f. coils so that they are in the center of their respective coils. Then set all of the audio i.f. slugs so that they are in the center of their respective coils.

Stagger Tuning

Stagger tuning is used to broaden the frequency response of the video i.f. It is not too difficult to accomplish without the use of a sweep generator if care is taken and every movement of the i.f. slug and the resultant effect upon the test pattern is noted.

First observe where or from what i.f. the audio is obtained. This will be the lowest video i.f. The slug will be set approximately in the center of the coil form. The last i.f. should be set for high frequency video information or at low i.f. frequency, i.e., the slug here will also be set at approximately the center of the coil form. It is therefore advisable to leave the last video i.f. and the video i.f. from which the audio is taken strictly alone. Between these stages the slugs in the video i.f. are set in gradual steps to cover the full coil length. If the audio is taken from any stage but the first video the stages remaining from the converter to this point are staggered in gradual steps from the high i.f. (slug out of coil) to the low i.f. (slug in center of coil), that stage immediately following the converter being set to high i.f. setting.

Now the set may be turned on and (Continued on page 168)

A SIMPLE ELE CTRONIC METRONOME

■HE metronome is a device well known to music students. In its - long-used mechanical form, it consists essentially of a small adjustable pendulum driven by clockwork, which audibly beats time. Aside from setting rhythm accurately for musicians and dancers, the metronome has found use in totally unrelated fields. Some applications include the beating of seconds in completely darkened photographic laboratories, timing of mechanisms when the eye cannot be transferred to a stop watch, timing of successive steps observed in chemical or biological processes when the sight is totally occupied with watching the process, marking time in physical training exercises, etc. An electronic metronome has the advantage of freedom from moving parts. It requires no winding and does not run down. Its volume may be adjusted to suit room conditions.

GUY DEXTER

The heart of any electronic metronome is a low-frequency oscillator or recurrent discharge circuit of some kind. The frequency of this oscillator is continuously variable over a small range. For simplicity, compactness, and good service, the circuit must be simple.

The author tested several circuits while developing the instrument described in this article. These included a multivibrator with two-stage audio amplifier, blocking oscillator and amplifier, gas-triode saw-tooth oscillator with amplifier, low-frequency resistance-tuned high-output oscillator, and neon-bulb oscillator with amplifier.

Results obtained with the neon-bulb oscillator driving a single-stage loudspeaker amplifier were good enough to justify turning away from the more complicated schemes. There has been a great deal of opposition to using neon-type relaxation oscillators in timing operations because of the instability of these units. However, the frequency of the neon circuit can be made sufficiently stable for metronome use by regulating the applied voltage and by using good-grade condensers in the circuit.

The final circuit (see Fig. 2) delivers readily-controlled beats ranging in rate from slightly less than 1 per second to 5 per second. This corresponds to a little under 60 to 300 per minute. The beats are delivered by a small loudspeaker and are "plops" which are not harsh to the ear. This metronome is small enough to be

Fig. 1. Complete metronome removed from cabinet. A neon lamp, 117N7, and OC3/VR105 comprise tube lineup. Controls are for frequency and volume. PM speaker delivers "beats."

Details for building a non-critical yet accurate marker for various laboratory and service jobs.

mounted in a midget radio or intercom cabinet and is inexpensive to build and operate. It is an ac.-d.c. device.

Circuit Features

The relaxation oscillator consists of resistors R_4 and R_5 , condenser C_5 , and neon lamp N_1 . Rheostat R_4 is the speed control. The highest beat rate is obtained at the minimum-resistance setting of R_4 . Resistor R_5 sets the high-frequency limit of oscillation by raising the lowest resistance value to which R_4 may be set. Condenser C_2 couples the oscillator output into the volume control, R_5 , and thence into the grid circuit of the amplifier. Adjustment of the volume control has no effect upon the frequency (speed). High-grade oil-impregnated, oil-filled .1 μ fd. condensers are used for C_2 and C_3 . For high stability of frequency, ordinary wax condensers must not be used.

The 117N7-GT tube supplies both the pentode amplifier and half-wave power supply rectifier. The cathode resistor (R_i) of the pentode is left unbypassed to obtain a small amount of degeneration. This serves to make the sound of the beats more agreeable to the ear. The pentode drives a 3-inch PM dynamic speaker which was found to deliver enough volume for an average size room. If louder beats are required, as for use in a large classroom

(Continued on page 143)

Fig. 2. Complete schematic diagram and parts list for the electronic metronome. $R_1 = -100 \text{ ohm}, 1 \text{ w. res.}$ $R_2 = -600 \text{ ohm}, 5 \text{ w. wirewound res.}$ $R_3 = -1 \text{ megohm pot.}$ $R_4 = -5 \text{ megohm pot.}$ $R_5 = -22 \text{ megohm, } \frac{1}{2} \text{ w. res.}$ $R_5 = -1500 \text{ ohm}, 2 \text{ w. res.}$ $C_2 = C_3 = -.1 \mu d_4, 400 \text{ v. oil-filled cond.}$ $C_2, C_3 = -.1 \mu d_4, 400 \text{ v. oil-filled cond.}$ $C_4, C_5 = -40/40 \mu d_4, 150/150 \text{ v. elec. cond.}$ $N_1 = \frac{1}{4} \text{ watt neon lamp (G. E. Type NE-48)}$ $S_1 = -S_{P.5.t.} \text{ sw. (on } R_3)$ $T_1 = -S_{P.5.t.} \text{ sw. (on } R_3)$ $T_1 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_1 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_1 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_2 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_1 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_2 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_3 = -S_{P.5.t.} \text{ sw. (on } R_4)$ $T_4 = -S_{P.5.t.} \text{ sw. (on } R_5)$ $T_4 = -S_{P.5.t.} \text{ sw. (on } R_5)$ $T_5 = -S_{P.5.t.} \text{ sw. (on } R_5)$ T





By JOHN B. LEDBETTER Eng., WKRC-TV

Profitable sidelines help stabilize a service business and provide a good economic cushion.

N THESE days of servicing complicated receivers and electronic - equipment, it is almost heresy to mention "sidelines" as a means of increasing shop income. The average service technician usually has his hands full keeping up with radio service alone. However, those little "odds and ends" often are welcome fill-ins and can often be built up to a point where employment of extra or full-time radio or appliance service technicians is justified. At the same time, these sidelines often attract new customers who might otherwise have taken their sets elsewhere for repairs.

There are many things which may be done to expand shop income. Here are a few: (1) sales and rental service of small p.a. systems, (2) installation of sound systems and intercom units in offices, large homes, schools, farms, and plants, (3) modernizing old-model a.c. receivers, p.a. amplifiers, intercoms, etc., (4) replacing electrodynamic speakers with new, more efficient PM units, (5) improving the performance of a.c.-d.c. receivers. (6) electrifying battery sets, (7) servicing small home appliances, (8) custom-building radio and television installations, (9) sales of records, record-players, and other attachments, and (10) operating a small recording studio in the shop.

Appliance Servicing

Small appliance servicing is probably one of the most effective ways of increasing shop repair work. Although not too profitable in itself, it *does* pave the way for more profitable work on radios, washers, refrigerators and other large household appliances. Each repair job on a small item provides an opportunity to suggest a replacement, a check on other appliances, an examination of radio sets in the home, or

for the SERVICE SHOP

the sale or repair of large appliances. Service can be expanded to include electric irons, toasters, waffle irons, hot-plates, heaters, lamps, fans, mixers, washing machines, vacuum cleaners, and even electric ranges, without requiring an expensive outlay of tools or replacement parts. Test instru-ments needed for most repairs need only include a volt-ohm-milliammeter with a 150 volt scale, an a.c. ammeter (0-20 amps.), and a simple a.c. test lamp. Anyone with any knowledge at all of simple a.c. circuits can repair small appliances; for washing machines, ranges, etc., where repair or replacement of motors, thermostats, etc., is necessary, a little more knowledge and preparation will be helpful.

Trouble in small appliances usually is limited to faulty connections in the plug or line cord, or breaks in the elements or wiring due to long use or rough handling. Permanent wave machines, hair dryers, ultraviolet lamps. etc., are subject to the above troubles; very seldom will trouble be found in their timing units.

Electric irons will perhaps need service more often than other small appliances. The usual difficulty is a faulty line cord or plug, or open or corroded elements or contacts. In automatic irons, the thermostat may be defective.

In electric mixers, fans, etc., motor trouble can be caused by dirty, worn brushes, lack of tension in the brush springs, or a dirty armature. Sometimes high mica between the armature segments prevents proper contact with the brushes. See that the motor is clean and properly lubricated. (Clean by disassembling and dipping each part in gasoline, or use a gasoline-saturated brush. Don't use carbon tetrachloride; it is corrosive). Worn take-up and thrust washers may be replaced, but for worn or "frozen" bearings, open arm-

ature, field coil, etc., the motor should be sent to a local automotive or electrical repair shop, returned to the factory, or replaced.

Vacuum cleaners have the same motor troubles just mentioned. In addition, the brush holders, fan, shaft, and intake tubes may be clogged with dirt, strings, wool nap, and other foreign matter picked up in cleaning. Usually the belt needs replacing. Check the switch, cord and plug as a safety measure.

Electric lamps (wall, desk, floor, etc.) are easy to service. Troubles include a defective switch, cord or plug, occasionally a bad socket. In fluorescent lamps, a bad lamp tube or faulty starter is usually encountered.

Electric ranges and washing machines are not difficult to repair. Defective heating elements in ranges can be replaced with exact factory replacements or universal elements; washing machine motors can be repaired at a local electric shop, returned to the factory, or replaced with a new unit. Wringer rollers, gears, and other worn parts can usually be obtained from the factory or from a local distributor. If the unit is so old that replacement parts are no longer available, the customer is a prospect for a new machine.

Another source of income is the *installation* of new electric ranges. Right now, you probably know several customers who are definite prospects for an electric range. You can do your "good deed" by recommending a particular dealer to the customer, and notifying the dealer of the customer's needs.

You may wonder how the radio service technician can profit by this. Here's how-many dealers do not employ regular installation men but call in outside help, and in many cases will be glad to pass this business along to you or your electrician associates. A good installation inevitably leads to future contacts with the customer and provides the opportunity to inquire about the condition of radios and other appliances in his home and in the homes of his neighbors. Don't overlook any of these opportunities. (A number of small, well-equipped service shops have carried the idea of sales, service, and installation a bit further by having an extra man (or men) specialize solely in work of this kind. Some have added refrigeration servicing as a profitable sideline). The Norwood Radio Service (see photograph on page 46) is a practical example of how sidelines may be added. This shop is not elaborate, neither does it let the grass grow in its show window. Note the sidelines offered: vacation and picnic portables, electric roasters, fans, irons, toasters (and for summer comfort, insect repellent lamps). Included are vacuum cleaners, washing machines, and, in the background, several console radios and an auto radio demonstrator-everything but the kitchen sink. Well, look again—the kitchen sink is there too! You may not want

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to go this far with sidelines, but you'll have to admit it's a good way to get customers into the shop, and it has paid off for this one.

Small P.A. Rentals

Many small amateur dance bands, schools, clubs, and other organizations need a low-power p.a. system occasionally but cannot afford the usual high-fidelity, high-power system. For these occasions, however, they are able to pay a nominal fee for a small 5- or 10-watt system. Old-model p.a. amplifiers can often be bought quite reasonably from large sound equipment companies and modernized, or if you prefer, you can build your own in spare time. A number of such rental systems in constant circulation will help pay for quite a few "incidentals" around the shop and, what is more important, will be valuable in promoting good-will and keeping your name before the public. (Along these lines, be sure your business name, address, and phone number are neatly painted or decaled on speakers, amplifiers and microphones). It is also good publicity to donate a p.a. system, receiver, record-player, etc., to civic organizations, business clubs, etc. It does not take long for cooperation of this sort to bring its rewards in increased sales, service, and business contacts. Many customers are connected with or own business concerns which could use intercom systems, paging systems, or sound equipment, or have equipment which is in need of repairs. In many cases the equipment is obsolete, or its effectiveness has been outgrown by expanding business. Situations of this sort are ripe for replacement units-make it your business to bring up these points. Sales can be helped along with a simple demonstrator like the one shown in Fig. 1. This set-up permits actual listening and comparison tests of amplifiers, low- and high-frequency speaker units, and record-changers; several different makes can be switched in for comparisons.

Small Recording Studio

A number of progressive service shops regularly make recordings of speeches, group functions, or special



Fig. 1. A simple demonstrator unit for merchandising amplifiers, low and high frequency speaker units, and recordchangers. It is a good sales stimulant.

occasions on inexpensive paper discs and pass them out to the participants. These make good "plugs" for the shop and often bring requests for more professional recordings from musicians, singers, and voice students who are interested in keeping permanent records of their progress. Small bands, school orchestras (and parents anxious to record junior's gurgling for posterity) help in making a small recording studio feasible. Many vocalists like to try new arrangements without the presence and attendant costs of a "live" orchestra. Special records and transcriptions (known commercially as "Synchro-discs") are available for this purpose. The vocal-ist and "canned" orchestra can be mixed very realistically.

Recording equipment, especially at the start, need not be too expensive. You can build your own mixer, preamplifiers, and recording amplifier, or you can use a rebuilt 5- or 10-watt p.a. amplifier if its quality is fairly good. (Continued on page 152)

Fig. 2. A special room for demonstrating the possibilities of custom installations set up by Altec-Lansing but equally adaptable to the smaller radio service shop.



The Variable Permeability Sweep Generator

By M. E. CLARK General Electric Company

HERE have been many approaches to the problem of wideband sweep generator design. Most of these have involved mechanically moving parts such as rotating or reciprocating variable condensers or a reciprocating disc moving near enough to a pancake coil to change its inductance as the disc moves. Such mechanically moving parts are subject to the usual problems of mechanical devices. Sliding surfaces wear and grow loose while flexed pieces fatigue and break. They are also subject to irregularities of motion when shocked or jarred as so often happens under ordinary working conditions in the laboratory or the service shop.

The obvious solution would be an electronic sweep such as the reactance tube type so commonly used for FM visual alignment work. In practice it turns out that it is extremely difficult to get sweeps of the width needed for TV work from the reactance tube sweep. The variable permeability sweep solves these problems by providing the needed width of sweep along with a total absence of moving parts.

As in any other design of an oscillator which is to be varied in frequency, it is necessary to vary either the inductance of the coil or the capacity of the condenser in the frequency determining network. Here we achieve the sweep by varying the inductance.

Variable Permeability Unit

As is generally known, the inductance of a choke falls off when the d.c. through it is increased. This is true because as the flux is increased in the iron the permeability of the iron is decreased and its effectiveness in increasing the inductance over that of an air core coil is lessened. Obviously, if this choke were a part of a tuned circuit the resonant frequency of the circuit would go up with saturation. By opening up a wide gap in the choke and putting a small piece of iron in that gap, it becomes possible to saturate the small piece of iron well in advance of the main part of the core because of its limited cross section. By using a piece of high quality radio frequency iron for the small slug and winding the oscillator coil on it we can cause a variation of inductance of the oscillator coil simply by running



Complete electronic control of sweep eliminates the many problems inherent in the often-used mechanical type systems.

d.c. through the winding of the original choke. See Fig. 2 for the development of this part which is the heart of the instrument under discussion.

Fig. 5 shows the final form of the variable permeability unit itself. L_1 is the choke or wide gap reactor, and L_2 is the high frequency oscillator coil. The small caps over the end of the laminated core serve merely to reduce losses by keeping r.f. out of the low frequency iron, and do not enter into the creation of the sweep.

When there is no d.c. through the choke coil, the iron slug in the tuned

Fig. 2. Inductance of oscillator coil is varied by saturating the r.f. iron slug.



circuit lowers the frequency of the oscillator by many megacycles over what it would be without the core. Now if the d.c. through the choke is increased the iron slug will gradually saturate and become less and less effective in increasing the coil inductance. Consequently, the oscillator will go up in frequency in much the same way as it would if the slug were withdrawn from the coil (as it is in many common iron-slug tuned circuits such as i.f. transformers).

Characteristics of the Unit

If frequency is plotted as the choke current is varied the resulting graph approximates a straight line as shown in Fig. 4, and the frequency is said to be linear with respect to the current. Note that if the current is reversed, the same frequencies are covered again. In other words, a saturated condition is reached regardless of which way the saturating flux happens to be flowing.

When a sine wave of voltage is applied to an inductance, the resulting current is also of a sine wave form, but shifted in phase. Since the wide gap choke coil (or wide gap reactor as it is often called) is nothing more than an inductance, the current through it will be a sine wave if a sine wave voltage is applied to it. The ra-

dio frequency will then be varied according to this same sine wave voltage because, as is shown in Fig. 4, the frequency varies the same way the current does. Saturation of the iron circuit would make this statement not strictly true, but saturation of the r.f. iron changes the over-all reluctance so slightly as to cause no trouble.

As also shown in Fig. 4, the frequency will repeat itself if the current is swung the same amount negative as it is positive. Under the condition where a sine wave is applied, this would cause the frequency to repeat four times. For example, going up toward 10 ma., it would pass through 205 mc. and it would pass back through 205 mc. on the way down toward zero. Then it would do the same thing on the negative half of the cycle, thus covering any one frequency (except at the extremities). four times. Two of these can be eliminated very simply by introducing a positive direct current of such a value as to prevent the total current from ever going in a negative direction. This is similar in principle to the negative bias on a class A amplifier which prevents the grid from ever going positive, but in this case the polarity happens to be reversed.

If the oscillator is turned off during the time when the frequency is sweeping back there will then remain only one repetition of any one frequency for each cycle of the sweeping voltage. This is referred to as blanking and is accomplished by biasing the oscillator grid so far negative that oscillations cease. The source and wave shape of the blanking voltage will be discussed later.

Now if the same voltage that is causing the frequency to sweep is applied to the horizontal amplifier of a scope, the spot will move across the screen at the same rate as the frequency is changing. By introducing a phasing control in the voltage applied to the scope it is possible to cause the spot to move not only at the same rate as the frequency, but to cause it also to be in step with the frequency at all times. We now have the basic unit upon which the variable permeability sweep was built.

A Practical Design

In applying the above principle to a practical design, it was found that a sweep oscillator could readily be made to cover the television channels in the frequency range of 165 to 220 mc. The lower channels and intermediate frequencies were then covered by taking the difference frequency between the swept oscillator and a c.w. oscillator. To effectively take this difference, a mixer was used in which one oscillator was fed to one grid of a dual triode. and the other oscillator to the second grid of the same tube. The plates were connected in parallel and the difference output taken directly from the plates into an attenuator. When the c.w. (or beat oscillator) is turned off the mixer becomes simply an ampli-



fier using one half of the tube to carry the swept oscillator output to the attenuator. Fig. 3 is a block diagram showing the above arrangement.

The high television band is covered by the direct output of the swept oscillator which covers from 165 to 220 mc. The low television band is covered by leaving the c.w. oscillator fixed at 275 mc., and tuning the swept oscillator over the high band again. A range of 275 minus 220 to 275 minus 165 is then covered. This is 55 to 110 mc., and covers the low television band and the FM band.

Next, the swept oscillator is left tuned to 220 mc. while the c.w. oscillator is varied from 220 to 275, thus yielding 275 minus 220 to 220 minus 220. This is a range of 55 to zero mc. Actually it is impossible to arrive close to zero output frequency because the oscillators lock together and there is then no beat output at all. With no special inter-oscillator shielding the generator works very well down to 4 mc. which allows it to safely cover 4.5





mc. intermediate frequency amplifiers. Differences rather than sums are used in all cases in order to remove the beating oscillators to frequencies above the beat output.

One of the most important characteristics of a good sweep generator is

Fig. 5. Rear view of generator. The frequency sweep unit is shown inside the shield.





Fig. 6. Schematic diagram of the sweep generator. Complete shielding is incorporated.

that the output remain constant while it is being swept across a band of frequencies. As might well be imagined the "Q" of the oscillator tuned circuit varies as the iron slug is saturated. This results in a variation of output of about 20% which is far too much for a laboratory quality instrument. Fortunately, this variation is about as linear with the current through the choke as is the frequency shift. This makes it practical to plate modulate by an approximately equal amount in the opposite direction. This is accomplished by inserting a modulation transformer in the "B plus" lead to

Fig. 7. Step attenuator. Capacity coupling, encountered when switches are used, is avoided.



the swept oscillator and modulating it with a little of the same voltage that causes the sweep.

Attenuator

In line with the constant output requirement it becomes absolutely necessary that there be no tuned circuits accidentally built into the unit which might cause resonant rises or take up some of the desired energy and cause the output to fall off for certain frequencies. This was taken into consideration in the design of the step attenuator by making all the leads as short as possible and inclosing the whole unit in a close fitting shield. The resistors were all strung out in a long line in order that capacitive and inductive coupling would be minimized as much as possible. Fig. 7 shows the construction and clearly illustrates the close fitting channel shaped shield. A rotary switch to replace the plug and jack arrangement would be very handy but the problem of decoupling one step from the next is difficult and expensive to solve.

A vernier attenuator was also desired, but whenever a potentiometer was placed anywhere in the circuit between the mixer and the step attenuator it caused the output to vary with frequency. It was found, however, that a low ohmage potentiometer could be placed at the end of the cable without causing any trouble provided that it matched the cable. While the vernier attenuator was put in the output box for a good electrical reason, it has since been found that it is in about the most convenient place possible as well.

Fig. 8 shows the output cable and the vernier attenuator just described along with a filter which will be mentioned later.

Leakage

If a signal generator, whether it be a sweep or not, cannot be cut down in output to match the weakest signal to be received it cannot do an adequate job of alignment. To meet this requirement it was necessary to resort to complete double shielding of the two oscillators, the mixer, and the lead which goes to the attenuator. Fig. 5 shows the double shields with the covers removed. All incoming leads were filtered between the shield cans and again inside the inner can to minimize leakage. Both condensers and one resistor of such a filter are shown in Fig. 9 and will be found on the schematic with the same symbol numbers. The two shields were connected together and grounded at only one point-where the output lead goes through. Fig. 9 illustrates the final solution of this problem. Note the double shielded cable passing over V_1 to the attenuator strip.

Power Supply

The power supply, to meet all requirements of this sweep design, is somewhat more complicated than one (Continued on page 94)

A Supermodulated PHONE TRANSMIT



Fig. 1. The 829B supermodulated transmitter. The small size of unit is shown in contrast to handset.

Constructional data on a small transmitter which uses the Taylor modulation system and is capable of handling as high as 90 watts of power input.

THE Taylor system of supermodulation has been a constant - attention attractor during the past year. Hams who have supermodulated rigs on the air have been peppered with questions regarding technical details of the circuit and method of adjustment. Receiving op-erators are intrigued by the characteristic audio punch of the signal, noticed even when the carrier seems weak. Authors of articles on the subject have been kept busy answering letters.

Considerable interest has been expressed in the use of supermodulation in mobile transmitters and especially in the design of a small transmitter which might be used interchangeably as a mobile rig or home unit. This article describes a transmitter which should satisfy the present demand. It is an inexpensive and compact unit.

The transmitter shown here uses an

By RUFUS P. TURNER, K6AI and JOHN W. GRAVES, WGMYR

829B tube in the final amplifier, one half serving as the power amplifier and the other as the positive modulator. All "supporting" stages, down to the speech amplifier and crystal oscillator, are self-contained. With the exception of the 829B, miniature tubes are used throughout. This transmitter has been developed especially for interchangeable mobile and home use and is the prototype of the forthcoming Taylor Model 100A. It is capable of 40 watts input at 500 volts on the 829B plates, and approximately 75 watts input at 750 plate volts. The input power has been increased safely to better than 90 watts.

The stage lineup is: 6AK6 20-meter crystal oscillator, 6AK6 10-meter doubler, 6AQ5 10-meter buffer-driver, 829B combination power amplifier and positive modulator, 6C4 speech amplifier, and 6AQ5 audio driver. A straight-through 10-meter buffer is used, since experimental data has shown that this type of driver gives better supermodulation efficiency and stability than a doubler. The number of stages was minimized by employing a 20-meter oscillator. However, 40meter crystals can be used if the builder is willing to add a 20-meter doubler to the stage lineup.

The complete circuit schematic is given in Fig. 3. Up to the output of

> Fig. 2. Rear chassis view of unit. The power terminal plug is seen on the left. The two vertically mounted jacks are J. and J₃ for PA and PM grid current indication. Single jack, J₁, is for microphone.





Fig. 3. Complete circuit diagram and parts list covering the 829B supermodulated transmitter.

the buffer stage, this arrangement is an entirely conventional capacitancecoupled exciter with separately-tuned tanks. The r. f. output from the buffer passes through the two coupling condensers, C_{14} and C_{15} in series, which together form a radio-frequency voltage divider. This divider maintains the proper ratio between r. f. grid voltages presented to the PA and PM sections of the 829B.

Audio voltages are presented to the final amplifier by the secondary winding of the special coupling transformer T_2 , condenser network C_{24} - C_{25} - C_{26} , and resistor R_{24} . Audio voltage at the center tap of the T_2 secondary is equal to one-half the voltage across the entire secondary winding.

The microphone input is intended for a Type F-1 single-button handsettype carbon microphone (See Fig. 1). However, any other single-button unit may be used. Excellent quality reports have been received when the F-1 unit has been employed. The 6C4 speech amplifier is a conventional stage with transformer input and resistance-capacitance output. The 6AQ5 audio driver tube is triode connected (screen tied to plate) to insure good audio output regulation and to provide about 1 watt of audio which is adequate for full modulation.

A swinging link, L_s , provides variable antenna coupling. This feature plus the antenna tuning condenser, C_{1s} , both are essential to peak performance of the supermodulation circuit. C_{1s} must be included in the circuit regardless of the impedance of the feeder cable employed.

A self-contained d.c. milliammeter, M_{1} , is switched successively across shunt resistors R_5 , R_9 , R_{14} , R_{15} , and R_{16} to read plate currents in the oscillator, doubler, buffer, power amplifier, and power modulator respectively. In addition, jack J_a is provided for plugging-in a low-range d.c. milliammeter to read PM grid current, and J_4 for PA grid current. Jack J_2 permits occasional checking of audio driver plate current. All three of these jacks may be dispensed with, since none is essential to tune-up operations.

The shunt resistor values must be selected to give the milliammeter a range of 0-100 ma. when reading oscillator, doubler, buffer, and PA plate currents, and 0-200 ma. when reading PM plate current. A 10-milliampere meter was used in the rig shown for the sole reason that it was immediately available. However, most builders will prefer to use the popular 0-1 ma. instrument. For the latter meter (with an internal movement resistance of 105 ohms), the shunts must have the following values: R_5 , R_9 , R_{14} , and R_{15} each 1.06 ohms (1 ohm will be close enough). R_{16} must be 0.527 ohm. The first four shunts may be made by winding the proper length of resistance wire on a ¼-inch diameter bakelite rod, or on the body of a burnedout carbon resistor. The last one (R_{16}) can consist of a small loop of resistance wire fastened to the two-terminal strip.

A 4-pair rear terminal strip receives the plug and cable from an external power supply. Two values of plate voltage (750 or 500, and 300), one

value of screen voltage (250), two values of grid bias voltage (-75 and -150 for 75 watts input) and (-50and -75 for 40 watts input), and 6.3 volt tube heater voltage must be supplied, as indicated on the appropriate terminal strip connections in Fig. 3. The plate and screen voltages may be derived from a single highvoltage power supply through a voltage divider. However, some means must be provided for reducing the final amplifier plate voltage during tune-up operations. This is very essential, since it is difficult to adjust the final amplifier properly while employing high plate voltage.

An important point in the transmitter design is the matter of placing the excitation taps on coils L_1 , L_2 , and L_3 . Note from Table 1 that these taps are taken near the *bottom* end of each coil. This is essential to proper impedance matching between one plate and the succeeding grid.

Construction

Mechanical arrangement of the transmitter may be noted from the photographs (Figs. 1, 2, 4, and 5) and from Fig. 1 which identifies the components.

The reader will notice the extensive use of interstage shields both on top of the chassis (See Figs. 1, 2, and 4) and below chassis (See Fig. 5). It is imperative to shield each of the exciter stages separately, as shown, and to shield the audio section from the exciter and final amplifier. Each tube in the exciter and audio sections has its own shield can.

A common ground is a must! Do not merely run the low end of a component to the chassis as a ground point, but make *all* returns in each stage to the *same* single point. Bond all of the separate stage-ground points together with a length of heavy bus wire. This is especially important in the 829B stage where multiple grounds are very apt to rob the transmitter almost completely of efficient supermodulation action.

The transmitter as shown is built on a metal chassis $11\frac{1}{4}$ " long, $2\frac{3}{4}$ " high, and 6" deep. This makes a compact assembly but the reader may deviate from these dimensions if necessary. To complete the shielding, a metal base is fastened to the bottom of the chassis, and the entire unit may be slipped into a metal case.

Power Supply

The power supplymust furnish 500 or 750 volts d. c. at 100 milliamperes full load, 300 volts d. c. at 30 ma., and 250 volts d. c. at 15 ma. The latter two voltages may be obtained from taps on a voltage divider. Additionally, two separate bias voltages (-75 and -150 for 75 watts input and -50 and -75 for 40 watts input), will be required. These may be obtained in the field conveniently from the nearest battery values, but at the home location probably will best be derived from an a.c.-operated bias supply.



Fig. 4. Top view of the transmitter showing the correct layout of parts. The oscillator, doubler, and buffer stages are shown at the left-front portion of the chassis. The audio section is immediately to the rear. The 829 stage occupies the entire right hand portion of the chassis. Note the adequate interstage shielding employed throughout.

6.3 volts a. c. or d. c. will be required for the tube heaters which draw a total of 3.6 amperes. For complete tube protection, the power supply should be arranged so that the bias voltages are applied automatically as soon as the heater voltage is switched on. The 829B screen voltage must never be applied ahead of the plate voltage.

Tuning Up

For best results, the following tuning operations must be followed closely.

1. Tune-up at a plate voltage not in excess of 500 v. to the 829B.

2. Swing the coupling coil (L_5) completely out of coil L_4 .

3. Unplug the microphone and turn the audio gain control, \mathcal{R}_{17} , to its "off" position.

4. (a) Switch the milliammeter, M_1 , to the oscillator stage and tune C_2 for plate current dip. Then, for maximum (Continued on page 104)

L ₁ (Osc.)	16 t. No. 18 en. on $3/4''$ d'am. form. Space to wind- ing length of $1/6''$. Tap 4th t. from "B-plus" end.
L ₂ (Doubler)	7 t. No. 18 en. on ³ / ₄ " diam. form. Space to winding length of 1". Tap 2nd t. from "B-plus" end.
\mathbf{L}_3 (Buffer)	5 t. No. 18 en. on 3/4" diam. form. Space to winding length of 5/8". Tap 2nd t. from "B-plus" end.
L, (Final)	Each half consists of 3 t. of $V_{6}^{\prime\prime\prime}$ copper tubing air- wound $1V_{2}^{\prime\prime\prime}$ in diam each in the same direction. Space each half to wind- ing length of 1". Spacing between halves $1V_{4}^{\prime\prime}$. Junc- tion of inside ends of the two halves forms center tap.
L ₅ (Ant. Coup.)	(Swinging Link) 3 t. $1/6''$ copper tubing airwound 11/2'' in diam. and spaced to winding length of $1/2''$. This coil is hinged so as to be swung between the halves of L_1 for var. an- tenna coupling.

Table 1. Coil winding data for L_1 through L_5 .

Fig. 5. Under chassis view of the transmitter. Note the use of baffle shields below the chassis as well as on top (Figs. 1 and 4). Rectangular cutout clears tuning condenser, C₁₆.





Compiled by KENNETH R. BOORD

T IS a pleasure this month to dedicate the ISW DEPARTMENT to

- - s.w. radio outlets in Brazil. The following current data on Brazil stations comes direct from Flavio Serrano, an *ISW DEPARTMENT* monitor in Rio de Janeiro:

The Tropical Band—4.785, ZYM6, Radio Ribamar, Sao Luiz, Maranhao, 0.5 kw.; 4.805, ZYS8, R. Difusora do Amazonas, Manaus, Amazonas, 5 kw.; 4.825, ZYE7, R. Educadora de Parnaiba, Parnaiba, Piaui, 0.5 kw.; 4.845, ZYU8, R. Difusora de Teresina, Teresina, Piaui, 0.5 kw.; 4.865, PRC5, Radio Clube do Para, Belem, Para, 2 kw.; 4.895, PRF6, Radio Bare, Manaus, Amazonas, 2.5 kw.

The 49 Meter Band-6.000, PRK5, R. Inconfidencia, Belo Horizonte, Minas Gerais, 5 kw.; 6.015, PRA8, R. Clube de Pernambuco, Recife, Pernambuco, 25 kw.; 6.085, ZYK2, R. Jornal do Commercio, Recife, Pernambuco, 16 kw.; 6.095, ZYB7,, R. Difusora Sao Paulo, Sao Paulo, Sao Paulo, 25 kw.; 6.105, R. Nacional, Rio de Janeiro, Distrito Federal, 50 kw.

The 31 Meter Band—9.500, PRL7, R. Nacional, Rio de Janeiro, Distrito Federal, 50 kw.; 9.505, PRL7, R. Nacional, Rio de Janeiro, Distrito Federal, 50 kw.; 9.565, ZYK3, R. Jornal do Commercio Recife, Pernambuco, 16 kw.; 9.61, ZYC8, R. Tamoio, Rio de Janeiro, Distrito Federal, 25 kw.; 9.77, PRL4, Ministerio da Educacao e Saude, Rio de Janeiro, Distrito Federal, 1 kw.

The 25 Meter Band-11.72, PRL8, R. Nacional, Rio de Janeiro, Distrito Federal, 50 kw.; 11.765, ZYB8, R. Difusora Sao Paulo, Sao Paulo, Sao Paulo, 25 kw.; 11.825, ZYK3, R. Jornal do Commercio, Recife, Parnambuco, 16 kw.; 11.95, PRL5, Ministerio da Educacao e Saude, Rio de Janeiro, Distrito Federal, 1 kw.

The 19 Meter Band—15.145, ZYK2, R. Jornal do Commercio, Recife, Pernambuco, 16 kw.; 15.155, ZYB9, R. Difusora Sao Paulo, Sao Paulo, Sao Paulo, 25 kw.; 15.165, ZYN7, Ceara Radio Clube, Fortaleza, Ceara, 10 kw.; 15.19, PRK9, R. Inconfidencia, Belo Horizonte, Minas Gerais, 5 kw.; 15.37, ZYC9, R. Tamoio, Rio de Janeiro, Distrito Federal, 25 kw.

The 16 Meter Band—17.85, PRL9, R. Nacional, Rio de Janeiro, Distrito Federal, 50 kw.

Construction Permits-For Tropical Band-Radio Poti, Natal, Rio Grande do Norte; Radio Brasil, Campinas, Sao Paulo, 1 kw. day, and 0.5 kw. night; R. Industrial de Juiz de Fora, Juiz de Fora, Minas Gerais, 0.5 kw.; R. Brasil Central, Goiania, Goias, 1 kw.; R. Quitandinha, Petropolis, Rio de Janeiro; R. A. Voz de Oeste, Cuiaba, Mato Grosso, 0.5 kw.; Government of the Territory of Guapore, Porto Velho, Guapore, 0.5 kw.; R. Relogio Federal, Rio de Janeiro, Distrito Federal. For Short-Wave Band—Radio Guaiba, Porto Alegre, Rio Grande do Sul.

Further comments by Mr. Serrano include-ZYE7, 4.825, has m.w. station on 1470 kc., call PRJ4, 500 w.; ZYU8, 4.845, has m.w. station on 1410 kc., call ZYQ3, 1 kw.; PRC5, 4.865, has m.w. station on 1450 kc., call PRC5, also, 500 w.; PRK5, 6.000, has m.w. station on 880 kc., call PRI3, 22 kw., to be increased to 50 kw.; PRA8, 6.015, noted on 6.016, has m.w. station 720 kc., call PRA8 also, 50 kw.; ZYK2, 6.085, effective radiated power is 64 kw., has m.w. outlet on 780 kc., 20 kw. day, and 10 kw. night, also FM stations (4); ZYB7, 6.095, has m.w. outlet on 960 kc., 5 kw., call PRF3, also FM station with call PRF3-FM; ZYN6, 6.105, has m.w. outlet on 1200 kc., 10 kw., call PRE9; PRL9, 6.147, has m.w. outlet on 980 kc., 50 kw., call PRE8, also on FM; PRL7, 9.50, 9.505, 9.72, see PRL9, 6.147; ZYK3, 9.565, see ZYK2, 6.085; ZYC8, 9.61, has m.w. outlet on 900 kc., 5 kw., call PRB7; PRL4, 9.77, has m.w. outlet on 800 kc., 25 kw., call PRA2 (power on m.w. to be increased to 50 kw.; will have new s.w. transmitter of 7.5 kw.

The sound effects department in action during a radio theater transmission at Radio Nacional in Rio de Janeiro, Brazil.



soon), this channel is inactive, station now radiates on 11.95; PRL8, 11.72, see PRL9, 6.147; ZYB8, 11.765, now inactive on this channel, owned by Radio Difusora Sao Paulo, but transmits programs from Radio Tupi, Sao Paulo, which has m.w. outlet on 1040 kc., call PRG2, 50 kw.; ZYK3, 11.825, see ZYK2, 6.085, as this channel is inactive; PRL5, 11.95, see PRL4, 9.77; ZYK2, 15.145, see ZYK2, 6.085; ZYB9, 15.155, noted on 15.156, transmits Radio Difusora Sao Paulo and Radio Tupi programs, also see ZYB7 and ZYB8; PRK9, 15.19, inactive, see ZYB8; PRK9, 15.19, inactive, see PRK5, 6.000; ZYC9, 15.37, is owned by Radio Tamoio but is operated by Radio Tupi, Rio de Janeiro, which has m.w. outlet on 1280 kc., PRG3, 50 kw.; PRL9, 17.85, inactive now, but is expected to be in use again soon see PRL9, 6.147.

PRN9, Distrito Federal Police Dept., 9.29, 5 kw., has daily transmission in Portuguese from 1700; lasts 30 minutes to two hours; programs consist of principal events and news of the Police Department; station maintains correspondence with overseas listeners; QRA is Departamento Federal de Saguranca Publica, Estacao de Radio, Rua da Relacao 55, Rio de Janeiro, Brazil.

Flashes just received from both Mr. Serrano and Hans Leven (of Sao Paulo) state that Radio Record, Sao Paulo, has just taken to the shortwaves-heard testing around 0600-2100 on 9.590 or 9.595; using m.w. call PRB9, which m.w. outlet, on 1000 kc., is one of the four m.w. stations in Brazil with 50 kw. power; QRA is Radio Record, Rua Quintino Bocaiuva 22, Sao Paulo, Brazil. Mr. Leven reports as new, Radio Difusora de Teresina, Teresina, State of Piaui, with call ZYU8, on (announced) 4.845 in parallel with m.w. ZYQ3 on 1370 kc. (m.w. outlet listed 1410 kc. by Mr. Serrano); heard with good signal to sign-off around 2000.

Mr. Leven says PRL5, 11.95, is scheduled weekdays 0400-1000, 1400-2030; Sunday 0700-2030; power 1 kw. (Continued on page 126)

⁽Note: Unless otherwise indicated, all time is expressed in American EST: add 5 hours for GCT. "News" refers to newscasts in the English language. In order to avoid confusion, the 24 hour clock has been used in designating the times of broadcasts. The hours from midnight until noon are shown as 0000 to 1200 while from 1 p.m. to midnight are shown as 1300 to 2400.) "The symbol "V" following a listed frequency indicates "varying." The station may operate either above or below the frequency given.

A HIGH EFFICIENCY TRIODE AMPLIFIER

TD'

MELVIN C. SPRINKLE

Peerless Electrical Products Div. Altec Lansing Corporation

Fig. 1. Peerless amplifier which is available in either the kit (10722) or completely assembled form (A-100A).

The Achilles' heel in their argument, however, has been the inefficiency of triodes. In spite of all the subjective arguments the fact remained that up to the present time triode quality was costly. In order to realize the full benefits of triodes it was necessary to operate with fixed bias-this meant either a higher voltage power supply with higher voltage condensers or a separate bias supply. Furthermore, because of the high bias required, transformer coupling into the final stage was necessary. Resistance coupled amplifier stages just don't put out high quality signals with enough amplitude to drive triodes. Both triode power supplies and/or transformer coupling are expensive.

On the theoretical side, triodes have had at least a strike and a half on them. Assuming proper power supply for both plate and bias, and good transformers, Terman states that the maximum theoretical efficiency of a Class A amplifier is 50%, and that practical amplifiers have efficiencies from 35% down to 20%. This meant that in order to get an output of 10 watts (developed plate-to-plate-not necessarily in the load) it was necessary to put 28.6 watts into a pair of tubes, and if the efficiency was of the order of 20% the input power had to be 50 watts. No wonder that triodes were limited to low output power! As a matter of fact the heads of many designers were scratched in the search for a high plate dissipation triode

Design details of a versatile audio amplifier incorporating a preamp and adjustable equalizing network for magnetic pickups.

without going into the expensive transmitting tubes. The development of Class AB operation was a partial solution as the efficiencies were raised. Terman gives efficiencies of 40% to 50% for this operation, but the tube manual in its typical operation data for 2A3 tubes says that the efficiency is 34.1% when the tubes are operated with fixed bias.

Class AB_1 operation of triodes has become accepted by audio enthusiasts who require high quality, but they have been in general very dubious about Class AB_2 . Up to the present time there has been some justification for their doubts. Recently, however, with the general availability of high quality transformers, operation of tubes up to, and possibly into, grid current becomes possible, with *no* more distortion than strict Class A operation, and with a prodigious increase in efficiency.

Recently the writer has become interested in a circuit which makes possible 18.6 watts output from a pair of 6A5 tubes at a distortion of 5% total harmonics! This represents an efficiency of 49.3%, a truly remarkable achievement. The amazing performance is made possible by two factors: (1) A good output transformer and (2) use of a cathode follower driver.

The output transformer is the heart of an amplifier. Almost invariably the output transformer is the only audio

transformer used, the other stages, with the possible exception of an occasional input transformer, being resistance-capacity coupled. The techniques for building wide range resistance coupled stages are well known and easily put into practice. If the results are not satisfactory, the builder can easily change values until the desired results are obtained. However, when one buys a transformer he must accept whatever the manufacturer puts into the unit. There are a few things that the user can do, but in general these have minor effects on over-all performance. For example, one cannot alter external circuits to improve efficiency-the ratio between primary power and the power delivered to the load.

It behooves the constructor, therefore, to buy the best possible transformer, considering the following points:

(a) Insertion loss. The useful power output of an amplifier is the power delivered to the load, not the power measured plate-to-plate on the primary side.

(b) Power output over frequency extremes. Vast numbers of amplifiers are built by audio enthusiasts who have absolutely no idea of how many watts their amplifiers deliver at 40 cycles and 10,000 cycles. Their conception of performance is based upon frequency response, which is a gain measurement, not a power measure-



ment. The factors which contribute to power output at frequency extremes are not controllable by the user.

(c) Frequency response. The frequency response of a transformer is a measure of its ability to transmit various frequencies; i.e., it is inherently a gain measurement. As pointed out above, the power output characteristic is often a more important property in output transformers, and it bears no relationship to the frequency response characteristic. On frequency response most manufacturers show response in the audible part of the spectrum, but what is more important is the response at frequencies above and below audibility. This response is of importance because in amplifiers employing inverse feedback, the response must be much broader than the range to be covered if the amplifier is to be stable.

(d) Concomitant with frequency response is phase shift. This is of importance if complex signals are to be transmitted without phase distortion and also if stability with inverse feedback is to be achieved. Phase shift, frequency response, and transient oscillation are most easily checked with square waves.

Most audio enthusiasts are familiar with conventional resistance or transformer coupling to Class A or AB1 amplifiers. These systems are quite satisfactory within certain limitations, but fail when Class AB₂ operation is approached. The fundamental reason is that when the grid of a tube is driven positive, instantaneously, there is a flow of electrons to the grid, which when they flow through the grid resistor, or even an interstage transformer. produce a voltage which is additive to the tube's bias. The effect is the same as the bias developed in an oscillator's grid leak due to grid current. In the usual Class AB₂ design a special transformer, called a driver transformer, is used, which is a stepdown transformer. In order to supply enough signal with a stepdown turns ratio, and also to supply power to the grids of the final when they draw grid current, it has been necessary to use a power tube as a driver.

Another factor that touches a sore spot in constructors is the cost of a good driver transformer. The cathode follower driver overcomes these limitations. It has been pointed out that a cathode follower is not a voltage amplifier, *i.e.*, the output voltage can approach but not exceed the input signal. However, and this is not as widely known, a cathode follower can be a power amplifier. The power in a resistor is given by the familiar formula $P = E^2/R$. Thus for a constant volt-

Fig. 2. Diagram of the Peerless 10722 amplifier. Circuit includes a preamplifier for magnetic phono pickups. With the addition of a special input transformer (circuit shown in lower left) the amplifier becomes suitable for use with a 30, 250, or 500 ohm microphone or other low-impedance, low level source. The equalizing network can be removed when used with microphone.

age, the power varies inversely as the resistance. Halving the resistance gives a 3 db. power increase for the same voltage. Thus a cathode follower can take in say 25 volts and in its output develop this same voltage (almost) across a much lower load, and thus give a power gain. By simple circuit manipulation a cathode follower driver stage can be direct-coupled to a final amplifier. It is possible to achieve excellent frequency response by the direct coupling and yet supply all the normal d.c. voltages to both stages. A low grid circuit impedance for both a.c. and d.c. is provided and there is sufficient power in the cathode follower to take care of grid current demands should operation in this region occur. The combination of two equally important factors-a good output transformer and a cathode follower driver-produces a high efficiency, high quality circuit that puts the triode on a more equal basis with beam power tubes.

This high efficiency circuit has been incorporated into a general purpose amplifier, specifically designed for high quality music reproduction, called the *Peerless* A-100A Amplifier.

With the development of this amplifier, the problem of low efficiency in triodes has been largely overcome. The unit has a maximum output of 20.3 watts. It uses two type 6A5 tubes in a newly developed high efficiency circuit which makes practical efficiencies of 49.3%. Moreover, and this is the crux of the matter, this power is useful secondary power, the implications of which were considered previously.

The frequency response of the amplifier is flat from 20 c.p.s. to 15,000 c.p.s. and is down only $\frac{1}{2}$ db. at 20,000 c.p.s.—an undetectable amount.

The distortion is remarkably low. Using the familiar 5% harmonic content standard, the power output is 18.6 watts at 100 and 400 c.p.s. At 40 c.p.s. the power output is 14.8 watts and at 5000 c.p.s. the power output is 18.3 watts. Using the intermodulation method with test frequencies of 40 and 2000 c.p.s., the power output at 8% intermodulation is 17 watts.

The schematic diagram is shown in Fig. 2, and it will be seen that the amplifier is entirely straightforward in design. The first stage uses a type 6J7 tube as a voltage amplifier. This stage is used as a preamplifier for the magnetic phono pickups so popular in high quality music systems. The plate circuit contains a series resistor-condenser combination, the values of which are adjusted to give the bass boost required for proper equalization of magnetic pickups. A feature of the amplifier is the fact that a chassis cutout is provided for installation of a Peerless K-221-Q 20/20 input transformer. With the transformer installed, the amplifier becomes suitable for use with a 30, 250, or 500 ohm microphone or other low impedance low level source, the equalization network being removed when a microphone is used.

Two inputs are provided, one for phono and the other for radio. A switch, S_i , selects the desired input and grounds the channel not in use to prevent crosstalk. In the phono channel a gain of 107 db. at 1000 cycles is available, which permits full power output from LP records. The radio channel has a gain of 84 db. which is more than adequate for any FM or AM radio tuner.

Following the radio-phono switch, is a four position, pi-type low-pass filter which is intended for record scratch reduction. The filter uses an inductance to give sharp cut-off so that the maximum usable recorded signal can be reproduced, and the higher noisefrequencies eliminated. The circuit must not be confused with resistorcondenser treble roll-off circuits and is quite different in performance. Four positions are provided, the characteristics of which are shown in Fig. 4.

In addition to the low pass filter, there is provided a continuously variable bass boost circuit. A resistorcondenser network is used to provide up to 10 db. of bass boost as shown in Fig. 4. A word of caution might be in order at this time. The bass boost control must be used with discretion, since the transformers provide considerable power at low frequencies.

The second stage uses a type 6J7 tube (V_2) , pentode connected to give high gain and very low distortion. A small portion of the cathode resistor is not bypassed as this is used for inverse feedback. The third stage (V_3) is a type 6J5 used in a split load or cathodyne phase inverter circuit. This type of circuit has several advantages among which are almost perfect balance in the output over wide frequency ranges, independence of changes in performance with tube replacement,

(Continued on page 134)



Fig. 3. Power output in watts vs. per-cent intermodulation distortion of amplifier.



Fig. 4. Amplifier response with low-pass filter and bass-boost circuits operating.



Fig. 5. Square wave patterns at various frequencies throughout amplifier range.

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Fig. 6. Under chassis view of the amplifier which may be built from a commercial kit.





By JACK TERRY

The experience gained in 42 years of radio is now being used to foster all segments of the industry.

HEN President Truman appointed George Edward Sterling a Commissioner of the Federal Communications Commission, he got not only a top-notch engineeradministrator; he got a radio amateur from Peake's Island, Maine. Sterling is the first and only ham to reach the bench of the FCC.

In 1908, at the age 14, young Sterling became interested in wireless. He talked his chums, Ray Hutchins and Al Ranger, into joining him in this fascinating scientific endeavor. The trio are well remembered by their teachers. One professor recalls, "they were smart as mustard—curious—particularly that George Sterling, but they didn't have the time to study ordinary academic things. Wireless was all that interested them and they would stay up all night combing the air waves with their home-made sets."

One of the trio reported: "We made almost everything we used. We had to buy our headphones though, and had a tough time getting the money. The local druggist, who also operated the Western Union office, helped us make wet batteries using salvaged carbon poles and zinc from old dry batteries. We made our spark gap out of an old copper lightning rod. That spark gap used to get so hot we were always worried about fire. Our spark transformer was wound by hand."

Reminiscing about those days, Sterling says, "When we started transmitting in 1908 we used the self-assigned call 'GS.' In those days we used a half inch spark coil in the transmitter.

"In 1912 the government licensed me as 1AE. Progress came in the form of a galena detector. And when we finally rigged up a variable coupler, the ultimate was in our grasp. The antenna was a big 4 wire flat-top hung between a couple of 65 foot poles out of the woods.

"The *Titanic* disaster was a firsthand thing to us. Advertised as an unsinkable ship, she rammed an iceberg on her maiden voyage on 1912 with the loss of over 1500 souls. We helped all we could with our wireless. We copied all the traffic—the pitifully small survivor list was an unforgettable event!"

In 1916 young Sterling went to sea as a pioneer radio operator. This career ended abruptly when the United States declared war on Germany. April, 1917, found his ship returning from South America. On arrival in New York, Sterling found orders to duty with his National Guard unit.

Among the first troops sent to France in the fall of 1917, his special radio talents were quickly recognized. He was transferred from the Infantry to the Signal Corps, commissioned a Second Lieutenant, and assigned as an instructor. It is interesting to note that these few months in the infantry mark his only separation from radio in the 42 years since 1908!

After World War I, the seafaring blood of his Maine ancestry was too strong to resist. It was inevitable that he should return to the merchant marine as a radio operator. For two years he traveled widely, following his profession. During this time he participated in salvage operations following an SOS from the *S.S. Suwide* and was paid \$300 salvage money according to Admiralty Law. He believes with pride that this is the only case on record of a radio operator being paid salvage money.

In 1922 the infant *Radio Corporation of America* invited young Sterling ashore as a Marine Radio Inspector.

A year later he began his long association with the enforcement of radio laws and regulations, joining the Bureau of Navigation, Department of Commerce, as a Radio Inspector.

Questioned as to his ham activities in those days, he replied, "I had a rig on top of the icebox in 1924. It was soon after I was married and space was at a premium in our apartment. That rig used a 5 watter—202—in a Hartley oscillator and its power supply used a chemical rectifier for which I borrowed all my wife's jelly glasses. They were later replaced by *AmRad* S tubes. The receiver was a *Reinartz* tuner. I had a lot of fun. In 1926-27, when I was writing the 'Radio Manual,' that same kitchen and ham setup were used to test out unfamiliar cir-(*Continued on page* 122)

Sterling at the controls. Present equipment at W3DF includes an SX-42 and SX-71 double-conversion unit (not shown in photo). The "ham shack" at W3DF. An HT-18 unit drives either a 600 w. phone or 400 w. c.w. rig. Transmitting gear is home-constructed.





Complete design characteristics of a tunable built-in antenna used for TV receivers.

By

ROBERT B. ALBRIGHT Eng. Dept., Philco Corporation

HE increased popularity of the built-in television antenna can be - attributed to various causes but the fact remains that most manufacturers have recognized the need for such a unit and are, consequently, equipping their receivers with selfcontained antennas.

In most cases the built-in antenna consists of a length of 300 ohm transmission line folded over inside the television receiver cabinet. This antenna may take the shape of an ordinary dipole or the transmission line may be fanned out from the center to form a sort of conical type. In all respects this antenna is essentially an untuned system, but one that can be used in high signal areas to receive a television signal. What are the disadvantages of such a system? The relatively small size of this built-in antenna makes for a very small signal capture area. This may be compared to a fisherman with a net. With a large net he will surely seine more fish than with a small net. Since the antenna is still

anuraceed for juently, h selfantenna Fig. 1. Over-all view a trans- of Philco's antenna he teleinto its being built into its video sets. rdinary may be form a espects ntuned

> more or less of a dipole, it will be difficult to secure a ghost-free picture in many urban areas. Again, since the antenna is essentially a dipole, its impedance at the feed point, that is, where the transmission line is attached to feed the front end of the receiver, will change radically above and below its resonant frequency.

two antennas, spaced at some angle less than 90 degrees from each other, permitting either antenna to be switched in separately to give some degree of directivity.

The tunable and matched built-in antenna system has several distinct advantages over the previously described types, for by providing a tun-(Continued on page 144)

Fig. 2. Both series and parallel resonances are used in Philco's built-in antennas.



May, 1950

Some built-in antenna systems use (Co





The GENESCOPE—A TV Test Instrument

By MILTON S. KIVER

A marker generator, an oscilloscope, a 400 cycle audio signal, and an FM and TV sweep generator are included in one test unit.

ADIO service technicians, together with radio engineers, long ago demonstrated by their eager acceptance of the multitester, that they prefer test instruments to be as compact and all-inclusive as possible. Thus, the vacuum-tube voltmeter, embracing within itself the functions of an ohmmeter, voltmeter, ammeter, and output meter is today easily the most popular and most important instrument in the service shop or the engineering laboratory. It is accurate, compact, and neat and can be used to perform any of the foregoing functions with a flip of a switch.

Due to this desire for speed and efficiency it was almost inevitable that an instrument like the Simpson Genescope should appear. Here, upon one chassis, is available an FM sweep generator, a television sweep generator, a 400-cycle audio signal, a marker generator, and an oscilloscope. In short, everything is provided for the proper alignment, signal tracing, and waveform servicing of FM and television receivers. This instrument, in conjunction with a vacuum-tube voltmeter, represents essentially all of the test and service equipment required by the modern television service shop.

The Genescope, shown in Figs. 1 and 8, is divided into three sections. The left-hand section contains a three-range r.f. generator, a crystal oscillator, and a 400-cycle audio oscillator. Frequencies generated extend from 3.3 mc. to 250 mc., divided into six ranges:

Band A—3.3-7.8 mc. (on fundamentals) and 6.6-15.6 mc. (on second harmonics).

Band B-15-38 mc. (on fundamentals) and 30-76 mc. (on second harmonics).

Band C—75-125 mc. on (fundamentals) and 150-250 mc. (on second harmonics).

The output of the r.f. generator may either be unmodulated or modulated by a 400-cycle note.

The center section of the Genescope contains a 3-inch oscilloscope and all its operating controls. In order to conserve space, the cathode-ray tube is mounted vertically. The pattern on the tube is then made visible to someone standing in front of the instrument by reflection from a tilted mirror mounted at the top of the cabinet. Provision is made for direct connection to the deflecting elements of the CRT by removing a small cover at the lower center of the panel. Fig. 1. The Genescope being used in the servicing of a television receiver chassis.

The right-hand section of the Genescope contains an FM signal generator, a 140 mc. fixed frequency oscillator, mixer, phasing, and blanking circuits. Frequency-swept signals can be obtained in two ranges extending from 2 to 120 mc. and from 140 to 260 mc. A sweep control permits adjustment of the frequency sweep from zero to over 15 megacycles.

A feature which is unique in this sweep generator is a blanking circuit which injects a negative pulse into the FM oscillator circuit in such a manner that oscillation is stopped during the return trace of the oscilloscope, thereby producing a base line and a single trace response curve on the oscilloscope screen. Thus, as the electron beam sweeps forward, from left to right, the response curve of the circuit under test is traced out. On the return trip, the beam would ordinarily trace back over the response curve produced on the forward sweep. This second tracing is not actually necessary since it provides the same information as the forward trace. Furthermore, there is generally sufficient unbalance existing in the circuit so that the second trace does not coincide at all points with the first trace, thereby producing two curves. See Fig. 3A. In the Genescope, this second trace is removed by stopping the oscillations of the sweep generator during this period. Within the scope, however, the beam is not similarly blanked out, but permitted instead to produce a zero voltage or reference base line. See Fig. 3B. The presence of this base line aids the technician to better orient the various values of the response curve and thereby tends to simplify and hasten the servicing and alignment process. The zero base line proves to be especially valuable for FM discriminator alignment since in this instance the linear portion of

the S-response curve should extend for equal distances above and below this level. See Fig. 3C.

Appropriate connecting cables are furnished with the instrument for making connections between the Genescope and the receiver under test. Also included is a high-frequency probe for signal tracing of high-frequency circuits; this connects to the vertical input jack of the oscilloscope section.

Applications

Any problem in a TV or FM receiver that requires an oscilloscope or an alignment generator can be handled by this instrument. Thus, consider the alignment of a television receiver. The usual order of adjustment is as follows:

- 1. Video i.f. traps
- 2. Video i.f. transformers
- 3. Sound discriminator
- 4. Sound i.f. transformers
- 5. The r.f. section

The video i.f. traps require adjustment at certain definite frequencieseach of which can be supplied by the AM generator in the Genescope. As an illustration, consider the typical video i.f. system shown in Fig. 5. The signal output lead from the Genescope would connect to point F of Fig. 5, while the ground lead of this cable would be attached to the receiver chassis. The vertical input cable (attached to the vertical amplifier of the oscilloscope) would be placed across the video detector load resistor-point A and ground in Fig. 5. The r.f. signal is amplitude modulated by the 400 cycle audio note in order that a varying voltage be developed at point A. The visual indication on the oscilloscope will be the 400 cycle sine wave.

Set the r.f. generator for a frequency of 19.75 mc. and adjust the core of L_{τ} for minimum height of the oscilloscope pattern. If the pattern disappears completely, increase the signal output of the generator until the exact minimum point can be observed on CRT screen. This trap circuit is now properly adjusted. In a similar manner, using the proper frequencies, traps L_{5} , L_{5} , and L_{9} , are adjusted, in each instance making certain that a definite minimum indication is observed on the screen. Particular attention should be given the



Fig. 2. Response curves for various portions of Philco's video i.f. See text.

21.25 mc. traps since these prevent the sound voltages of the signal being received from reaching the image tube screen.

Coils L_2 , L_1 , L_6 , L_8 , and L_{11} provide the video response characteristic for this i.f. system and each coil is peaked at its own frequency for maximum indication of the 400 cycle sine-wave pattern on the CRT screen.

Although the accuracy of the AM generator is designed to exceed 1 percent, even greater accuracy is required when adjusting television receivers. For this reason the Genescope is provided with a crystal standard having an accuracy of .05 per-cent or better. By using this standard and a special scale provided for the AM generator (in addition to its regular scale), frequencies having crystal accuracy may be obtained at any point within the range of the AM generator. This calibrating procedure would be applied to the trap and i.f. peaking frequencies just mentioned, enabling the service technician to adjust these coils to their exact resonant frequency.

After all of the trap and i.f. adjustments have been completed, we are ready to view the over-all response curve of the video i.f. system. Do not molest the connecting cables between the Genescope and the receiver. These remain intact as set for the previous



Fig. 3. (A) The response pattern of a video i.f. system as it normally appears. Note the double trace. (B) The same pattern as shown in A with the blanking voltage on. (C) A discriminator response with the special blanking voltage on.



Fig. 4. A video i.f. response curve showing position of marker pip at the point where the video carrier should appear.

adjustments. Merely switch the FM generator to the appropriate range and adjust the sweeping range for a value of 6 to 7 mc. Rotate the phasing control until the two traces visi-







Fig. 6. A conventional sound i.f. system found in many television receivers.



Fig. 7. (A) Discriminator "S" curve before the blanking voltage is applied. (B) The same curve after the blanking voltage is applied. (C) Appearance of the 400 cycle wave, as indicated in text, signifies that discriminator secondary requires some retouching.

ble on the sereen blend as closely into each other as possible and then advance the blanking control until only a single trace is seen on the screen accompanied by a straight base line extending across the screen. If the video system has been properly adjusted, the curve shown in Fig. 3B will be obtained.

Since the video response extends for 3 to 4 mc. (depending upon the receiver), it is impossible to judge from a simple viewing of the curve itself whether the carrier and end frequencies are where they should be. Hence, identification markers are required and these are supplied by the AM generator. This generator is switched to the unmodulated position and set to the i.f. frequency of the video carrier, say 25.75 mc. A pip should appear halfway up the right-hand slope of the response curve. See Fig. 4. To determine the extent of the curve, the frequency of the AM generator is changed until the marker pip appears along the left-hand slope of the response curve. The frequency of the pip can now be read from the scale of the AM generator.

Care should be exercised that the marker pip is kept as low as possible,

Fig. 8. A front panel view of the Genescope, Simpson's new television service instrument.



otherwise the response curve will be seriously distorted and it will be difficult to establish the exact position of the trace. In this respect a separate attenuator permits the operator to adjust the amplitude of the marker pip independently of the FM sweep signal.

The foregoing alignment considered only a stagger-tuned video i.f. system. Equally good results can be obtained with video i.f. systems using complex coupling. Du Mont, Stromberg-Carlson, and Philco television receivers are good examples of such systems. For these systems, the alignment begins with the last i.f. stage and proceeds back to the mixer. To start, the vertical input cable of the Genescope is connected across the video detector output while the cable from the FM sweep generator is connected between grid and ground of the last video i.f. amplifier. The response curve for the tuned circuit between this last i.f. stage and the video detector will appear on the scope screen. The shape of this curve should be compared with the curve recommended by the manufacturer (in his service manual) for this i.f. stage. If a difference exists, the proper adjustments should be made to bring the response curve into line. Marker frequencies, if desired, can again be supplied by using the AM generator portion of the Genescope.

In *Philco* receivers, the vertical input cable of the oscilloscope is left across the video detector output while the sweep generator output is moved back in turn from the last video i.f. stage (the 3rd video i.f. amplifier) to the 2nd video i.f. stage to the 1st stage and finally to the mixer. In the latter position, the response curve of the entire video i.f. system is obtained on the scope screen. The response patterns for each position of the sweep generator are shown in Fig. 2, together with various identifying frequencies.

In Du Mont and other television receivers, the manufacturer recommends that the response pattern of each i.f. stage be considered by itself. To enable the service technician to do this, a high-frequency probe with a self-contained detector is provided. The probe output leads are connected to the vertical input terminals of the Genescope oscilloscope. Now, by connecting the sweep generator leads to the control grid of one i.f. amplifier tube and placing the probe at the control grid of the following i.f. amplifier tube, we will obtain a response curve of the tuned circuit contained between these stages. This procedure can be followed in any tuned circuit in the video i.f. system. In some instances where the stage gain is low, it may be necessary to feed the signal through two stages to produce a usable indication on the scope screen. In this instance, the tuned circuits between both stages would have to be adjusted together.

(Continued on page 108)

NBFM And Frequency Shift Keying Unit For Self-Excited Oscillator

By R. W. JONES, W6EDG Chief Communications Tech., USN

F THE many different types of narrow-band frequency modulation circuits none have been described for amateur application using the Miller effect of a vacuum tube. The unit to be described, using the Miller effect of a 6SL7GT, gives good narrow-band frequency modulation with the added feature of frequency shift keying for eleven meter operation.

The input capacity of a vacuum tube depends not only on the actual gridcathode capacity but will vary with changes in the amplification factor or transconductance of the tube. The change in input capacity with a change in amplification factor is known as the "Miller Effect."

The input capacity of a triode tube with a resistive plate load is calculated from the formula:

 $C_{in} = C_{Gk} + C_{Gp} (A+1)$ Where: $C_{in} = input$ capacity in $\mu\mu fd$.

- $C_{Gk} = \operatorname{grid}$ cathode capacity in μµfd.
 - $C_{q_p} =$ grid-plate capacity in μµfd.
 - \boldsymbol{A} = gain of triode vacuum tube amplifier.

Note that changing the value of Awill change the input capacity of the triode vacuum tube. The amplification of a vacuum tube can be varied by two methods, changing the grid bias or changing the plate voltage. In the application of Miller effect to be described in this article I believe that the plate voltage variation method is the better of the two methods because of better isolation of the audio frequency voltages from the radio frequency energy in the oscillator to be frequency modulated.

The complete diagram is given in Fig. 1. The 6SF5 is a conventional audio amplifier stage to amplify the audio voltage from the crystal mike. The heart of the unit is the 6SL7GT. a twin triode with separate cathode connections. Note that the two plates of the triodes are fed plate voltage through a common plate resistor, R_{6} .

Any change in plate current to the first section of the 6SL7GT will change the plate voltage on the second section because of the change in voltage drop

The unit, consisting of two tubes as shown, is mounted directly above tuning condenser of author's v.f.o. It is also bolted to the front panel.



Design is based on the "Miller Effect" of a 6SL7GT. Provides good NBFM and FSK over eleven meter band.

across the plate resistor with changes in plate current. This change in plate voltage in the second section of the 6SL7GT will cause a change in amplification of the second section. Referring again to the formula, a change in the gain A of a triode stage will cause a change in input capacity. The input circuit of the second section of the 6SL7GT is across the grid in-

ductance of the oscillator circuit. Changes in input capacity of the 6SL7GT will cause a change in frequency of the oscillator. The input circuit of the second section of the 6SL7GT is, in reality, then a variable condenser across the grid inductance of the self-excited oscillator. The variable condenser changes capacity with (Continued on page 98)

Diagram of NBFM and FSK unit. The entire design is based around the 6SL7 tube.



Experimental Nagnetic-Type D.C.-A.C. Signal Inverter

SGOOD . GLOBE CORD

JAMES KAUKE

LECTROMECHANICAL vibrators or choppers usually are employed as tubeless devices to convert small d.c. voltages to a.c. for measurement or control purposes. A familiar example of the use of such devices is in the amplification of d.c. voltages through a conventional a.c. amplifier. In this application, the small d.c. voltage is turned into a.c. by a vibrator or chopper somewhat similar to an auto radio vibrator. The a.c. is then applied to the input terminals of a regular a.c. amplifier (which can be made much more stable than most d.c. amplifiers). Finally, the output of the amplifier is rectified, and a high d.c. voltage (proportional to the original small voltage) is obtained which may be readily measured or used to actuate a husky d.c. relay.

Electromechanical inverters have proven unsatisfactory to most amateur experimenters because of the presence of moving parts which are subject to sparking, sticking, wearing, noise generation, unsteady output amplitude, and frequency shift. Highly refined vibrators designed for stable instrument use are priced out of the reach of the ordinary experimenter.

By making use of the saturable reactor principle, a simple low-level d.c.to-a.c. inverter can be constructed without moving parts or tubes. The operating principle of the saturable reactor is quite simple. In its simplest form, this device consists essentially of two coils wound transformerfashion on a suitable iron core. An alternating current is passed through one winding in series with a load circuit. The d.c. is applied to the other winding. The inductance (and reactFig. 1. (Top) A simple, two-transformer type saturable reactor for d.c. to a.c. inversion. This unit was built for installation on audio amplifier chassis to be used for measuring d.c. millivolts. Connections are made to the terminal strip lugs and to load resistor. See Fig. 3A for circuit. The a.c. supply transformer was eliminated as the necessary voltage was obtained from the amplifier filaments.

Fig. 2. (Left) A simple magnetic d.c. amplifier. This unit uses the circuit shown in Fig. 3B. A cable, connected to the terminal strip point, connects with input and output terminals and the power line which is installed in the unit's mounting case.

Complete details on a new gadget. This saturable reactor type of tubeless magnetic amplifier can be used for instrumentation and control purposes.

ance) of the a.c. winding changes in accordance with the amount of d.c. flowing through the d.c. winding. As the inductance changes, so does the alternating current flowing through the a.c. coil and load circuit. In this way, d.c. changes are converted readily into a.c. changes. This explanation applies to the most rudimentary type. Most commercial saturable reactors sufficiently sensitive for instrumentation and control are more complicated, but the basic operating principle is the same.

If the core of a saturable reactor has the proper magnetic properties, a relatively large alternating current in one winding can be controlled by a small d.c. in the other winding. This is a type of amplification and is the principle of the *magnetic amplifier*, a device which is assuming increased importance in electronic applications. Saturable reactor devices require only a straight a.c. power supply, an advantage when working for simplicity and compactness.

Like the instrument-type vibrators, commercial high-quality saturable reactors and magnetic amplifier-transformers are priced beyond the budgets of amateur experimenters. This article describes a simple, inexpensive saturable reactor-type d.c.-to-a.c. inverter which can be constructed by the experimenter. Many radio technicians will have the necessary parts already on hand. Aside from the applications listed by the author, the simple inverter can be used for numerous other purposes which will suggest themselves to the ingenious experimenter.

Simple D.C.-A.C. Inverter

A saturable reactor can be made, as shown in Fig. 3A, with two UTC Type S0-2 "Subouncer" transformers, T_2 and T_3 . These are subminiature transformers having cores with high nickel content which enables them to be used much more successfully as saturable reactors than conventional transformers. The secondaries are connected in parallel to handle the a.c. The d.c. signal input voltage is applied to the primary windings which are connected in series *bucking* to eliminate any induced a.c. from the d.c. input terminals. The d.c. input ter-

minals have no particular polarity. The a.c. operating voltage is supplied by a 6.3 volt filament transformer, T_1 . This may be the lowest-current unit available, since the drain is not severe. The a.c. output signal voltage is developed across a 400 ohm load resistor, and may be applied to any high-impedance device, such as an a.c. vacuum tube voltmeter or a.c. amplifier. The signal output voltage will be as steady as the a.c. operating voltage.

A steady value of 0.1 volt a.c. appears across the 400 ohm resistor when no d.c. is applied to the input terminals. This may be reduced to zero by means of an a.c. bucking voltage of the same value applied in series with the resistor, or by means of a bucking voltage applied in series with one of the output leads. If this initial voltage is of no consequence in a particular application, it may be neglected entirely.

The operating curve for this d.c.a.c. inverter is given in Fig. 5B. It will be observed from this curve that the simple inverter is not an amplifier, the a.c. output voltage being lower than the applied d.c. voltage. However, this is not important, since the chief application of the simple inverter is ahead of a standard highgain a.c. amplifier, as shown in Fig. 4. In an application of this sort, the small a.c. output voltage of the inverter is applied to the amplifier input terminals. The 6.3 volt operating voltage may be obtained from the filament circuit in the amplifier, and transformer T_1 dispensed with, if desired. The a.c. output of the amplifier is rectified by means of a crystal or selenium rectifier and used to actuate a d.c. milliammeter or husky d.c. relay. A suitable amplifier to use is the familiar 6SJ7-6J5-6V6 (or 6L6) combination. With the inverter-amplifier setup, small d.c. voltages in the millivolt region or less can be measured or used for performing control operations.

It will be noted from the curve, Fig. 5B, that the inverter response is not linear. The shape of the curve is caused by the magnetic properties of the transformer core material. This nonlinearity will occasion little difficulty, however, since a special scale or calibration curve can be drawn for the meter in the amplifier output circuit when the device is to be used for measurements. When the combination is used for control purposes, fixedpoint operation ordinarily is desired, and curve shape is of little consequence. Input resistance of the d.c. circuit is 500 ohms. The dotted curve in Fig. 5B shows the d.c. milliamperes drain over the operating voltage range. While the 500 ohm input is a relatively low value, it is considerably higher than the resistance of common d.c. millivoltmeters ordinarily employed.

Fig. 1 shows an assembled simple inverter with transformers T_2 and T_3 and the 400 ohm load resistor. The



Fig. 3. (A) A simple d.c.-a.c. inverter. (B) Magnetic d.c. amplifier ("d.c. transformer").

6.3 volt transformer was eliminated in this model, since the inverter was built for installation in an amplifier having 6.3 volts available at the tube filament terminals.

Amplifier-Type Inverter

When more voltage output is desired than is supplied by the simple inverter, the 400 ohm output resistor may be replaced by the primary winding of a step-up transformer. This arrangement gives a simple magnetic amplifier, the gain of which is determined by the step-up turns ratio of the output transformer.

Fig. 3B shows such a circuit. Transformer T_4 steps up the output voltage which then is turned back into d.c. by the 1N34 rectifier. The circuit has been referred to frequently as a "d.c. transformer" because it, in effect, steps up a small d.c. input voltage to a higher value through a transformer. If a.c. output is desired, the 1N34 may be omitted and output taken directly from the T_4 secondary.

This circuit permits use of a d.c. vacuum tube voltmeter (connected to the d.c. output terminals of the inverter circuit) to check small d.c. voltages. The curve in Fig. 5A shows operating characteristics of the circuit of Fig. 3B. Note that the voltage gain varies from 5 (for a d.c. input of 0.1 v.) to 11.8 (for a d.c. input of 0.8 v.). The average gain over the d.c. input range of 0.1 to 1.1 v, is 10.4. At d.c. input values lower than 0.1 v., gain falls off rapidly because of the magnetic properties of the T_2 and T_3 cores. This may be improved, along with the gain at all other points, by using a transformer with higher turns ratio than the one specified in Fig. 3B for



Fig. 4. The d.c. amplifier setup.

 $T_{\rm 4}$. Inexpensive line-to-grid transformers, for example, are available with turns ratios of 60 to 1 and higher.

A steady d.c. voltage of approximately 1.2 v. appears at the d.c. output terminals when no d.c. is applied to the input terminals. This may be reduced to zero by means of a bucking voltage of the same value applied between the amplifier (inverter) output and the input of the high-impedance load device.

It should be noted that the circuit given in Fig. 3B is strictly a voltage amplifier. It gives neither current nor power gain. For that reason, it must be operated only into a high impedance, such as a d.c. vacuum tube voltmeter or relay-tube grid circuit.

Fig. 2 shows an amplifier type of inverter completely assembled.

Voltage Variation and Impedance

In each of the circuits given, variations of the 6.3 volt a.c. supply voltage will affect the a.c. output voltage value. However, because of the stepdown ratio of the 6.3 volt transformer, higher line-voltage excursions can be tolerated than in other circuits. When the inverter is followed by an (Continued on page 155)

Fig. 5. (A) Operating characteristics of the magnetic d.c. amplifier shown in Fig. 3B. (B) Operating characteristics of simple inverter shown schematically in Fig. 3A.



Fig. 1. Commercially-built fan type antenna designed and constructed by Channel Master Corp. engineers.



DEVELOPMENT of FAN TYPE TV ANTENNA

Design of an antenna which maintains the good broadband performance of conical antennas yet provides better high-band gain characteristics.

HE past year has seen the phenomenal emergence of the coni-- cal type as the predominant suburban and fringe area antenna. Its characteristics are superior to the folded or straight dipole type of antenna from the standpoint of gain and of bandwidth. The use of separate high and low sections can be eliminated in most cases. However, this antenna does have several drawbacks. The first is the poor response brought about by its drooping gain characteristic on the

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high end of the high band from Channels 10 to 13. (See Fig. 2.) This is accompanied by a splitting of the major lobe into two large side lobes in the horizontal polar pattern. (See Figs. 3A and 3B.) This means that there is both loss in gain on these channels and a high susceptivity to ghosts due to the splitting lobe. The sharp slopes in the gain curve indicate that in a channel having a 6 megacycle bandwidth, there can be differences in gain in an antenna of 2 to 3 db. These sharp slopes

Fig. 2. A typical curve showing the gain of a conventional conical antenna.



and non-uniform gains introduce distortion to video components of the telecast signal.

About six months ago work was undertaken at the *Channel Master Laboratories* on the design of an antenna which would match or improve the broadband and gain characteristics of the conical antenna without the deterioration of the horizontal polar pattern on the high end of the high band and the resultant drooping gain characteristic.

A brief discussion of field strength standards would help to clarify some of the points which will follow. Field strength contours from transmitters are generally plotted in microvolts-per-meter. Contours are generally shown for 5000 µv., 1000 µv., 500 µv. and 100 μ v. This means specifically that on the 100 μ v. contours, a wire one meter long will intercept that number of microvolts. Fig. 4 shows three contours for a theoretical telecasting setup. For the purpose of this illustration, both Channel 2 and Channel 13 are telecasting from the same point with exactly the same amount of power. A tuned dipole for Channel 2 is roughly two meters long and a tuned dipole for Channel 13 is about half a meter long. This means that at the 100 µv. contour where the one meter wire reference is placed, the Channel 2 dipole will give approximately 200 μ v. since it is twice as long as the reference and the Channel 13 dipole will give 50 μ v. since it is only half as long as the reference wire. It can be seen that although both tuned dipoles are placed on the same field strength contour, one will give four times as much signal as the other.

* Chief Electronic Engineer.
 † Sales Manager.

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In other words, in areas of equal signal strength, tuned dipoles for low band channels will supply stronger signals to the set than tuned dipoles for high band channels. Moreover, transmission line losses are greater on the high band, and in general, receiver front ends are less efficient on the high band. The chief problem then is to design antennas which can supply more energy on high band channels than the conventional half-wave dipole or dipole and reflector combinations.

Since this is the problem, more energy must be intercepted by the antenna. This means that the high band dipole must be longer than a half-wave. On the average, a half-wave dipole for the low band is three times as long as the half-wave dipole for the high band. This means that three times as much signal would be intercepted on the high band with this dipole. However, since on the high band this low band dipole is three half-waves long, it is operating on the third harmonic at high band frequencies and the resultant lobe pattern is a clover leaf somewhat similar to Fig. 3B. This split lobe means that the antenna is insensitive to "head on" reception and that if one major lobe is pointed toward the station, the other is highly receptive to ghosts and stray pickup. (It is assumed that rear lobes will be reduced by the use of reflectors.) The problem then is to keep the antenna length but to "focus" the energy into one lobe. This is accomplished by arranging the antenna so that it forms a "V" (Fig. 3C). The "V" type antenna gains efficiency as the legs of the "V" become a greater number of wavelengths. The higher the channel, the greater the gain and the narrower the lobe.

However, when this is done, low band efficiency decreases. The reason for this is that veeing forward of the elements reduces the space aperture or area of interception. In addition to this, the nulls at 90 degrees are not complete. (Fig. 3D.)

At this point in the discussion, it is necessary to deviate slightly to consider the evolution and theory of a conical element. What is ultimately accomplished is the incorporation of these two principles, the "V" type element and the conical element into one broadband antenna. The conical element was developed out of the need to create a dipole antenna which would operate efficiently over a wide range of frequencies and also present a relatively uniform impedance over this range of frequencies. A dipole can be considered a special case of the parallel wire transmission line. As is commonly known, the characteristic impedance of such a line depends upon the diameter of the conductors and the spacing between them and is independent of frequency. If such a line is opened and bent so that the two conductors are formed into right angles opposite to each other and each conductor is cut a quarter-wavelength from the bend, a half-wave dipole is obtained. Before the bend, while the



Fig. 3. (A) Horizontal pattern of half-wave dipole cut for Channel 2. (B) Pattern of same antenna when operated on its fourth harmonic or Channel 13. (C) Pattern of antenna shown in B when the elements are bent into the form of a "V". (D) pattern of this same "V" antenna when it is used on Channel 2.





wires are still parallel to each other, voltages and currents in each conductor are equal and opposite as are the fields surrounding these conductors (Continued on page 132)



Fig. 5. Polar pattern of fan antenna.



Nac's RADIO SERVICE SHOP

By JOHN T. FRYE

T WAS really too nice an afternoon to work. The whole early - May day had been a sneak preview of late June. Outside the open door of Mac's Radio Service Shop the warm sun beat down on the concrete sidewalk that had been washed clean by a shower the night before and sent shimmering heat waves rising through the quiet air. Only the ever-busy, worm-stretching robins seemed to feel industrious.

Inside the service shop that noon hour there certainly was no out-oftune display of ambition. Mac himself occupied the high stool that was precariously balanced on two legs so that he could lean back against the wall of the service department. Across from him Barney, the apprentice, sat on the service bench with his long legs doubled up and his arms clasped about them so as to form a rest for his chin. "Hey, Mac," the boy finally drawled. "Uh huh," Mac acknowledged.

"A guy down at the radio club last night asked me what kind of a system of troubleshooting you use. He looked at me kind of funny when I said I didn't know. What system do you use?"

The stool came down on four legs with a thud as Mac straightened up. "That darned question always makes me mad!" he exploded. "A lot of people can't rest until they get any form of activity boiled down to a neatlylabelled little 'system.' The next time he asks, just tell him we use the looking-and-thinking system.'

"Don't you believe in using a regular method in running down trouble?'

"Depends on what you mean by 'regular method.' If you mean working logically from observed symptoms MAC'S "SYSTEM"

to likely causes of trouble, I certainly do: but if you mean a religious and exclusive following of the voltagechecking system, the resistance-measurement system, the signal injection system, the signal-tracing system, or the touching-the-grids system-which I now see labelled with the hoitytoity name of 'circuit disturbance method'-the answer is a flat and emphatic No."

"You don't use any of these?"

"I use all of 'em—and a lot of other 'systems' besides. As far as I am concerned, I need everything I can lay my hands and brains on in running down trouble in a receiver. When an intermittent and I go to the mat, it is a fight to the finish with no holds barred. Those devils employ every trick they can hatch up to keep you from finding what ails them, and if you fight fair by following some sort of electronic Marquis of Queensberry rules, they'll tie you up in knots.

"Just look at how dirty they play: a mica condenser, that ought to be like Caesar's wife, above suspicion, will become leaky just often enough to escape an investigation until you have exhausted every other possibil-ity. And then there are those sets that behave like angels on the bench but start cutting out just as soon as you have them all bolted down in the cabinet. When you are dealing with fiendish cases like those, it is no time to be hobbled with a single system. You need every method of diagnosis you can conjure up and a strong dash of inspiration besides."

Barney drew a careful bead on the doorknob with the solder gun as he asked, "You say you need to use everything?"

"That's right. Take the eyes: when I turn on a set, the first thing I do is to look at the rectifier to see if it glows that gorgeous shade that indicates gas or if the plates turn a cherry red because of a shorted filter condenser.

"The nose comes in mighty handy, too. One whiff of the inside of a cabinet in which a power transformer has been too hot is all a service technician needs to tell him that he had better do some checking before he turns that set on.'

"And I suppose you use the master's touch to fix them," Barney suggested.

"Modesty forbids my saying Yes," Mac replied with a grin; "but the sense of touch is really a very useful service tool. It will locate a stonecold metal tube with an open filament or spot a leaky filter condenser that is running a temperature and should be replaced; moreover, a finger probed lightly around the rim of a speaker with a rubbing voice coil will quickly tell you where it is rubbing and which way it should be moved to free it.

"I don't think I need to go into any illustrations of the hundred and one things your ears can tell you about a sick set. If you just know how to listen, it will babble all its symptoms to you. I must admit, though," he said musingly, "that I have never been able to figure out a way to use the sense of taste to help fix radios. I wish I could."

"If you forget how the polarity of a bias cell goes, you might try touching a couple of leads from it to your tongue," was Barney's bright sugges-tion. "I think the positive lead—or maybe it's the negative one-is supposed to taste salty."

"I'll remember that valuable suggestion," Mac promised dryly. "Seriously, though, all of the well-known methods of running down trouble have good qualities, and I want you to be perfectly at home with any of them. Quite likely any one of them, if followed carefully, would eventually spot a trouble; but the service technician can make much better time if he has all of the methods at his finger tips and can change freely from one to the other as he tracks down a receiver fault. Each of the systems should be looked on as a service tool. Just as you pick up first one tool and then another when you are working on a mechanical job, so you should be able to switch from one system to another when you are cornering an electronic trouble."

"Let's have a 'ferinstance,' " Barney suggested.

Well, for instance, suppose you turn on a set and start checking the voltage. You find voltage at the rectifier cathode and the plate of the output tube, but there is none on the i.f. tube plate. It is a good idea to turn the set off at once because of the possibility that there may be a short somewhere that will burn up a resistor-perhaps inside the i.f.-or do

(Continued on page 82)



By HERB MATTHEWS Fig. 1. Over-all view of audio amplifier built by the author. The unit incorporates many of the design features covered in text. Schematic diagram appears on page 72.

Part 2. Frequency characteristics of various RC tone control circuits used in high-quality audio amplifiers. The concluding article of the series.

CONE control" is an ill-used phrase which has been em-- ployed to describe a multitude of functions and hide a multitude of sins. Tone controls may range from a simple capacitive roll-off, intended to reduce noise and distortion, to a complex *LCR* network designed to make every music lover his own conductor with complete control over the tonal balance of the orchestra. Neither of these extremes are befitting a high fidelity system. (The term "high fidelity" is used in its pure sense with blatant disregard for its corrupted connotation.)

Tone control, or equalization, may be considered in two distinct classifications: equalization to offset deficiencies (which are generally fixed) in the reproducing system and equalization to offset deficiencies in the original program material. Unless this distinction is recognized it is impossible to design or operate tone controls intelligently.

Deficiencies in the reproducing system may include anything from a phonograph pickup to a human ear. The effect of loudness variations of the ear was covered in Part 1; any other deficiencies in the ear need not affect the audio designer. If a listener has a 10 db. peak in hearing perception in the range of 1000 to 3000 c.p.s., there is no point in trying to change his tonal perspective in the reproducing system because all sounds will be natural to him with the frequency distortion to which he has become conditioned. If his hearing falls off rapidly above 5000 c.p.s., however, do not expect him to appreciate the full tonal beauty of your wide-range system.

As was illustrated in a preceding article, variations in the loudspeaker system are something which frequently occur and for which compensation may be introduced. Variations in amplifier transformers may also be compensated for by equalization. However, poor frequency response in a transformer is usually indicative of other deficiencies and though the frequency response may be flattened out by means of equalization, other distortions cannot be removed so simply. Where high quality is desired, the best obtainable audio transformers should be used.

The loss of high frequencies in amplifier circuits because of shunt capacities and Miller Effects can readily be compensated by simple treble boost circuits. Low frequency loss is primarily a result of undersized coupling and screen and cathode bypass con-





densers. These components are relatively cheap and you can generally afford to use larger sizes than those specified by many designers with beneficial results. Small losses in various stages of an amplifier can multiply to surprising proportions; when an amplifier is found to have inadequate frequency response it will usually be due to more than one cause.

Deficiencies in input sources such as phonograph pickups, microphones, and radio tuners can usually be made up but not always as readily as those previously discussed. When compensating for high frequency losses it must be remembered that a rising high frequency curve emphasizes distortion. A distorted wave (Fig. 2A) having its high frequency components (harmonic distortion) emphasized would look greatly distorted (Fig. 2B) if observed on an oscilloscope. If, however, a wave of the form of Fig. 2A were passed through a system with high frequency loss (Fig. 6Å) and then through a corrective equalizer, its harmonic distortion would remain unchanged. There is, therefore, a limit to the benefit accruing from treble equalization and the best means of achieving real high fidelity is the simple expedient of having every part of the system flat. (This, of course, is a bit more difficult than it may sound.) Corrective equalization is a very useful device, but its proper use must be based on careful consideration of all factors.

The acoustics of the room, then, should also be considered as a property of the reproducing system. A heavily carpeted room with drapery and upholstered furniture will absorb high frequency energy to an abnormal degree. This will not only cause an



Equivalent circuits (B) show basic operating principle of all RC tone control systems. Z_1 (equal to R_p) remains constant while Z_2 (the acc-resistance of C_2 and R_t in parallel) varies with the frequency.







Fig. 5



Fig. 4 ceivers.

acoustic attenuation, particularly of the higher frequencies, but will also shorten the reverberation time in the treble region, effecting a marked loss of brilliance. For this reason high frequency boost may be desirable in this type of room; conversely a treble droop may be required in a room with large, smooth-surfaced areas. Other characteristics of the room may affect the response in other portions of the audio spectrum. For example, a long room may act as a resonant air column and produce a peak at certain low frequencies which will result in boominess.

After the reproducing system has been made as flat as possible, the system may still sound unbalanced because nothing has yet been done to equalize variations in the program source. These variations should actually be compensated for in a recording or broadcast studio control room, but unfortunately there are not only differences of opinion as to what correct balance should be, but there are also different recording standards. Low frequency turnover points vary from 250 to 1000 c.p.s. and high frequency pre-emphasis varies considerably among different recording studios. In many broadcast studios it is the practice to place bass instruments abnormally close to the microphone as a means of compensating for poor low frequency response in most home re-

These external variations in frequency response may change every time a radio program source changes or a different record is played. For this reason "tone controls" intended to compensate for these changes should have great flexibility and should be convenient to use. The author firmly believes that the only panel variable tone controls provided should be for

(A) 08 RELATIVE (B) (C) (C) 20K 20 100 10K FREQUENCY IN C.P.S.

Fig. 6

compensation of program differences and that the deficiencies peculiar to the reproducing system should be taken care of by fixed pre-set equalizers.

Now that the need for equalization and the philosophy by which it should be applied has been explained, actual tone control circuits may be discussed. The most familiar circuit is the treble cut-off tone control found in most commercial radios. It may be a single condenser switched across the audio section (Figs. 3A and 6A) or a volume control type of rheostat which may be used to connect it into the circuit gradually (Figs. 4 and 6B). In the circuit of Fig. 3A it will be observed that the impedance (a.c. resistance) of C_2 will vary with frequency as follows:

At all low frequencies up to a certain point, P, (Fig. 6A) which shall be called the take-off point, the impedance will be higher than the total impedance across the tube, R_t . $R_t =$ $(R_1+R_2)/R_1R_2$. Don't be afraid of the algebra. All the formula means is that the resistance R_t , is equal to the total resistance of the two parallel resistors shown in Fig. 3A. It will be necessary to understand the behavior of R_t in all circuits to be presented in order to apply them properly to any given existing circuit. If you are not yet used to the mathematical form of presentation, now is as good a time as any to familiarize yourself with it. The more you understand about the behavior of radio circuits the more you can do with them.

To continue with the action of a treble cut-off; as the impedance of C_2 approaches that of R_i , C_2 will begin to absorb power and a voltage dividing network will be operating. From this point on, the amplification will halve as the frequency doubles. This is equivalent to saying that the frequency response of this circuit drops at the rate of 6 db. per octave. The limit which can be accomplished with any single stage of RC tone control of any form is 6 db. per octave. Because of the shunting action of R_1 across C_2 the theoretical maximum of 6 db. per octave is never reached.

When a resistor is connected in series with C_2 (Fig. 4), it limits the maximum effect that the condenser can have and further reduces the slope. As shown in Fig. 6B a variable resistance can make the curve move vertically to any desired degree within the 6 db. per octave limit. Changing the value of C_2 will make the curve move horizontally as desired. Doubling C_2 would move the take-off point one octave (a frequency ratio of 2 to 1) toward the low frequency end of the curve. Using just C_2 it can be seen that the treble range would be seriously affected, but if a limiting resistor is inserted into the circuit the entire frequency spectrum is easily controlled. Fig. 5 shows a practical circuit used in the amplifier illustrated in Fig. 1 to produce bass boost with the response of Fig. 7A. It will be noticed

that a pentode tube is specified. The reason for this is that the plate resistance of a triode is so low that the triode operates at constant voltage and the voltage dividing network described above will not operate. With a pentode the plate resistance is quite high and a constant current condition exists. Where a triode is used, the circuit of Fig. 8 will produce approximately the same frequency response; however, the insertion loss of the frequency controlling network will be somewhat greater.

It will be helpful to review the various points of Fig. 6C so that if a different curve were desired by the reader he might readily make changes in the circuit constants to suit his requirements. The point labeled A is the point at which the low frequency response falls off. As in any other re-sistance coupled circuit this point is controlled by the size of the coupling and bypass condensers, C_1 , C_3 , C_4 of Fig. 5. Satisfactory formulas for computing these values for any desired frequency will be found in the "Sylvania Technical Manual" (page 25, 7th edition). The point P is the same as previously discussed and is controlled by the size of C_2 . For any given value of C_2 the location of point B is determined by the size of limiting resistor R_{3} . If point P were to be moved down one octave by doubling C_2 , point B would also move down one octave. To move point B back up one octave halve the resistance of R_3 . This effectively increases the length of the slope P-B, and therefore increases the total boost and the insertion loss. See Fig. 7.

It may be seen that any response desired may readily be obtained along a 6 db. per octave slope using simple RC circuits. Obviously a smaller rise than 6 db. per octave might be required and this may readily be accomplished by placing a shunt resistor (R_{5} Fig. 10) across C_2 or by making R_3 (Fig. 5) large. By making $R_{\rm s}$ variable a simple variable bass boost results. By using two stages of bass boost a 12 db. per octave slope may be approached and by using a dual pot to serve as R_5 in each stage any desired boost from 0 to nearly 12 db. per octave results. (Fig. 10.) Constants for this circuit are the same as for Fig. 5 $\,$ with R_{δ} equal to 1 megohm.

It might be required that a complex curve be produced for low frequency equalization. Combinations of various boosts such as those shown in Fig. 7 could be added together by making the constants of the two stages different. The permissible combinations are almost infinite. Another arrangement which permits the take-off point of the curve to be moved about at will by means of a panel control is shown in Fig. 9. For the constants given, rotation of R_2 will produce any of the curves shown in Fig. 11 including intermediate values. The operation of this novel circuit is easily explained. With R_2 set nearest R_1 both condensers are shorted to ground and no boost will result. With R_2 at the center po-

May, 1950



sition the two condensers are effectively in series and therefore their effective capacity is .008 μ fd., resulting in a 6 db. per octave boost starting at about 2000 c.p.s. This characteristic is ideal for aural compensation at low volume levels. As the control moves toward the other end the effective capacity approaches .05 μ fd., resulting in a 6 db. per octave boost rising from about 300 c.p.s. This control has been found to be very effective.

All of the circuits discussed so far are of the low-pass type. The curves produced all slope downwards and the only factor that determines whether they are low or high frequency controls is the size of C_2 . Let us now turn our attention to circuits whose curves slope in the other direction, the high pass type of equalizer. Fig. 12 illustrates a filter so simple that you may not recognize it, although you may have unwittingly designed it into some of your own equipment.

 C_1 is the basic high pass filter. Its characteristic slope is also 6 db. per octave and its low frequency effect may be limited by a shunt resistor across it. Fig. 14 illustrates a simple









high pass circuit which will suffice for almost all treble boost circuits. The







Fig. 14

condenser is made variable in order to move the take-off point horizontally along the frequency spectrum. The value shown will allow wide flexibility with the rise starting anywhere between 1000 c.p.s. and 6000 c.p.s. The

Fig. 15



selector switch is included only to show that this circuit will work from any source. Fig. 13 is a variation of Fig. 14 which shows a potentiometer connected to permit either treble boost or treble cut. The take-off points for the boost and cut are entirely independent and are controlled by the value of the condensers employed. Many variations of these basic illustrations may present themselves. As an example of the type of thing which can be done Fig. 15 illustrates a combination of Fig. 14 and Fig. 8 which will give a combined high and low boost with a minimum insertion loss.

It is realized that some readers may not wish to take the time and trouble to work out circuits which suit their individual requirements precisely and therefore Fig. 16 has been included to show a completely self-contained preamplifier which may be used to feed an existing power amplifier or other equipment which the reader may possess. This unit includes a preamplifier for a magnetic phonograph pickup such as the GE, Pickering, or Clarkstan. The switch included in the phono preamp section is designed to equalize the pickup for various low frequency turnover characteristics as indicated. The treble control incorporated is the same as shown in Fig. 13 and discussed above. The bass control will provide a complex characteristic which will give approximately 24 db. maximum boost at 50 c.p.s. With the controls set at the flat position this unit is capable of uniform frequency response of ± 1 db. over the entire range of 20-20,000 c.p.s.

It is suggested that the power supply section be kept as far from the input section as possible to minimize hum difficulties. The large amount of bass boost called for presents some difficulties in keeping hum out of the circuit and care should be exercised. Grounds have been grouped on the schematic and you are strongly urged to follow the recommended grouping in your wiring. There should be only one place on the chassis where the electrical ground and the chassis are mechanically connected. The best way to stay out of ground loop difficulty (the worst source of hum in home built equipment) is to forget that the chassis can be a conductor and plan all ground returns so that each stage is wired to its own ground point and separate ground points are tied together by a deliberately wired connection. Do not use the chassis as part of the filament circuit. If your power transformer has no center tapped filament winding, connect one side of the filament circuit to ground at one point only.

Prefer to shield hum generating sources rather than the audio wiring. Shielded pair on filament wiring and a metallic shield around the power supply section generally accomplish this satisfactorily. By not shielding audio wiring itself, the circuit capacity is reduced and 20,000 c.p.s. response is readily attained. -30-

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0-30 Mc. — mechanical driven inductive sweep. Husky 110V. 60 cycle power transformer operated — step type output attenuator with 10,000 to 1 range — high output on all ranges — band switching for each range — vernier driven main calibrated dial with over 45 inches of calibration — vernier driven calibrated indicator marker tuning. Large grey crackle cabinet 161% "x 10%" x 7-3/16". Phase control for single trace adjustment. Uses three high frequency triodes plus 5Y3 rectifier — split stator tuning condensers for greater efficiency and accuracy at high frequencies — this Heathkit is complete and adequate for every alignment need and is supplied with every part — cabinet — calibrated panel — all coils and condensers wound, calibrated and adjusted. Tubes, transformer, test leads — every part with instruction manual for assembly and use. Actually three instruments in one — TV sweep generator — TV AM generator and TV marker indicator.


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UNIVERSAL TEST SPEAKER KIT

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Now a bench 6 Volt power supply kit for all auto radio testing. Supplies 5 - $7\frac{1}{2}$ Volts at 10 Amperes continuous or 15 Amperes intermittent. A well fil-tered rugged power supply uses heavy duty selenium rectifier, choke input filter with 4,000 MFD of electrolytic filter. 0 - 15 Volt meter indicates output. Out-put variable in eight steps Excellent for put variable in eight steps. Excellent for demonstrating auto radios. Ideal for servicing — can be lowered to find sticky vibrators or stepped up to equivalent of generator overload — easily constructed in less than two hours. Complete in every respect. Shipping Wt., 18 lbs.



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The husky 110 V. cased power transformer is conservatively rated for long life. The illuminated six inch slide rule dial is accurately calibrated for DX reception. Enjoy the pleasure of assembling your own fine home receiver. Has tone, volume, tuning and phono-radio controls. Chassis size $2!4'' \times 7'' \times 12!5''$ Comes complete with all parts including quality output transformer to 3.4 ohm voice coil, tubes, instruction manual, etc. (less speaker). Shipping Wt., 10 lbs. No. BR-1 Receiver \$19.50.



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Multi-TV Receive Operation

One solution to the problem of antenna-laden rooftops is the installation of a "master" unit.

By DANIEL LERNER TV Field Service Engr. Philco Corporation

URING the past few years the number of television receiver owners has increased at an amazing pace. It is predicted that there will be nine million television receiver owners by the end of 1950. With this marked increase comes the problem of providing enough space for outdoor antennas in crowded urban areas. In many locations in New York the larger apartment buildings are beginning to resemble houses with a fringe on top. The probable solution lies in a central antenna system which can be used to feed a large number of receivers.

Many such systems have been tried, but probably one of the most efficient is the recently developed unit called the "Antenaplex." This system is adaptable for apartment houses, hotels, institutions, and for numerous other applications.

Fundamentally, this system includes either a separate antenna for each channel in one particular area, or merely a separate antenna section for each channel on one mast. Of course the antenna system used depends upon the type of reception in the particular location.

The signals from the individual antenna are fed to a master amplifier



Fig. 1. "Antenaplex" multiple antenna system.

Fig. 2. Multiple antenna system used in department store and dealer installations.









Fig. 3. Simple system for feeding two, three, or four sets by multi-pad arrangement.

which increases the strength of the signal and provides the proper isolation to prevent receiver interaction. The outputs, of a necessarily low impedance to prevent noise and interference pickup, are fed over coaxial cables to room outlets. See Fig. 1.

Another type of system used in some department store, dealer, and school applications is shown in Fig. 2. Here we have a series of 6J6 tubes used as isolation amplifiers. The master antenna is fed to the grids of the 6J6 in parallel, with one triode for each side of the balanced 300 ohm line. It should be noted that a 300 ohm match is maintained at both the grid and plate sides of the triode. With this type of unit quite a bit of loss, in the order of 60%, should be expected but good isolation is effected nevertheless. Thus receiver local oscillator interaction is kept to a minimum. By using a booster amplifier before the signal is fed to this system, the effects of this loss may be minimized.

A simple system for feeding more than one receiver is a multi-pad arrangement as shown in Fig. 3. With this system a large amount of initial signal is required as the available signal decreases in proportion with the number of receivers fed. In all cases, however, the 300 ohm impedance is kept constant. $-\overline{30}$ -

SET OWNERSHIP IN FOREIGN COUNTRIES

FROM the Consumers Merchandise Branch of the United States Department of Commerce comes new and additional data on set ownership in various foreign countries, as reported by U. S. legations, embassies, and consulates.

Austria has an estimated 1,280,000 radio receivers in operation, as of November, 1949, of which 64,000 are unlicensed. Approximately 356,000 of the sets are equipped to receive short-wave broadcasts, and an estimated 3 persons listen to each receiver.

An estimated 25,000 radio receivers are in use in Burma of which 7000 were manufactured prior to 1939. Radios of United Kingdom manufacture dominate the Burmese radio market, followed by the United States and the Netherlands. Import controls on radio equipment have been liberalized since September 1949, resulting in increased imports.

Canadian manufacturers sold a total of 460,649 units during the first nine months of 1949 at a total value of \$33,326,189, an increase over the same period in 1948.

Imports during January-September 1949 amounted to 41,416 sets of which 39,974 sets were designated as "sets imported under special conditions". Exports of radio sets in the same period totaled 25,210 units.

Canadian production of radio receiving tubes in the first nine months of 1949 totaled 3,363,410 units, while for the corresponding periodin 1948, 3,241,-728 tubes were produced. Imports of receiving tubes during the January-September 1949 period aggregated 914,-352 units as compared with 759,122 in the same period of 1948. Imports of radio tube parts during January-September 1949 were valued at \$391,862 as compared with \$216,004 in the like period of 1948.

The report from Finland indicates an estimated 727,000 receivers in use as of December 31, 1949, of which 662,000 were licensed. About 71 per-cent are equipped to receive short-wave broadcasts. Approximately 3.6 persons listen to each set.

An estimated 42,000 radio receivers were in use in Iceland as of December 15, 1949. Approximately 65 per-cent of the sets in operation were Philips (Netherlands and the United Kingdom), 7 per-cent were Marconi (United Kingdom), 4 per-cent Telefunken (Germany), 7 per-cent other European brands, and 17 per-cent United States manufacture.

An estimated 80,000 radio receivers are in use in Iran, most of which were manufactured after 1945. About 60 percent of the sets are of United States manufacture, 30 per-cent British, and the remainder Dutch and Swiss.

Because of the shortage of clectric power, particularly in the provinces, an estimated 50 per-cent of the radios currently imported are battery operated. The life expectancy of batteries by the time they reach the Iranian market is reported to be as little as two months.

Some European exporters extend three months' credit and are reported to be contemplating extending credit up to six and nine months. It is reported that Iranian importers complain of the continued failure of United States exporters to extend credit.

-30-



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VETERANS: CREI training is available under the G.I. Bill. For most veterans July 25, 1951 is the deadline—so act now!

SAMPLE REE LESSON

Send for "The Orthicon and Image Orthicon" which de-scribes the development, theory and operation of the orthicon and image orthicon TV camera tubes.





Mac's Service Shop (Continued from page 68)

some other damage. You simply switch over to the ohmmeter and use the resistance-checking method to find out if your trouble is an open i.f. transformer primary, an open plate-isolating resistor, a shorted bypass condenser, or maybe a shorted tube."

"That makes sense," Barney observed judiciously.

"Another fault with being married to one system," Mac went on, "is that it tends to kill a guy's originality. A good service technician ought to be constantly trying to find better ways of doing his work. Just last week I had a convincing demonstration of that. For twenty years I've been trying to cook up a really good way to service those sets that are plagued with a rustling sound. You know what I mean: the kind of sound you get when an i.f. winding is opening up or an output primary is going out or a grid cap solder connection is poor or you have a noisy plate resistor.

"Those rustling sounds are produced by small, instantaneous current fluctuations. They are too quick to show on a meter, and you can't go probing around with an ohmmeter without running the risk of temporarily 'healing' the trouble you are trying to spot by running the ohmmeter current through the coils or resistors. Even a signal tracer cannot be used to advantage unless it is of extremely high gain and very, very quiet.

"The other day, when I had one of these sets, I suddenly got the bright idea of putting the vertical input of the scope across the units I suspected while the sweep was working on a low frequency and drawing out about a three-inch horizontal line. With the input across the output transformer, every little rustle would shoot up a little hair-like filament from the line. These little offshoots kept right on showing up as I moved back through the receiver until I reached the grid of the audio stage. On the plate I saw them; on the grid I didn't. When the tube was yanked, they were still there for a few seconds until the rectifier cathode started to cool-it was an a.c.d.c. set---so the tube was not at fault. When the mica r.f. bypass was cut loose from the plate, no little spikes were present across the resistor. That

condenser was leaky *only* when a certain critical voltage was across it. It checked OK with the battery voltages ordinarily used with an ohmmeter.

"From now on I intend to use the scope to track down all of these 'rustlers.' If my mind hadn't been burdened with the idea that a scope was only good for aligning, waveform analysis, and distortion-spotting, I might have discovered this much sooner. The less hidebound your service technique is, the more likely you are to discover new and better methods of doing your work.

"Really it all goes back to how you feel about radio servicing. Personall/, the thing I like about servicing is that it is a thinking business and cannot be easily systemized. The problems you encounter are far too complex and varied and changing to allow you to settle into a dull routine; and, by the same token, they are not to be solved with the greatest efficiency and dispatch by the application of any neat little universal formula or procedure. I am confident that the next set I put on the bench-and incidentally it is high time we started putting some sets on the bench-will be a little different in some detail or other from any I have worked on before-say," he broke off, "what the heck are you trying to do with that solder gun?"

"Just following your advice and working out a new and better method of servicing," Barney explained blandly. "It always gripes me when I want to see if a voice coil is open and I have to unsolder the output transformer secondary from across the coil before I can test it. About half the time I lose the flexible voice coil lead out of the terminal and have a tough time working it back. Now I've got a better way. When you pull the trigger on this gun anyways near the output transformer, you get a strong hum from the speaker if the transformer secondary and the voice coil are both OK. It must be a.c. induction from the gun transformer."

"My boy, I am proud of you!" Mac beamed. "You are really using that old flame-colored noggin. That makes me feel good."

"That's not what makes you feel good," Barney observed.

"What does, then?"

"Getting the 'systems' out of your system," the boy said with a knowing grin as he slid from the bench. -30-



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The twenty-five chapters into which the book is divided cover the entire field of ham communications from the basic fundamentals to the latest techniques in single-sideband telephony.

Since few persons in the field need any introduction to the "Handbook" suffice it to say that this edition is the "best ever."

"RADIO TECHNOLOGY" by Ernest J. Vogt. Published by *Pitman Publishing Corporation*, New York. 542 pages. Price \$6.00.

The publishers of this book have dubbed it a "one volume reference library" and they are not far wrong.

Covering telegraphy, telephony, television, transcription, and facsimile in addition to radio, this text packs an amazing amount of basic information into its 500 plus pages.

Designed as an interim handbook which fills the gap between elementary radio texts and books at an engineering level, the author has managed to present his subject matter without resorting to complicated mathematical formulas. What little math is required in the text is fully explained in a chapter devoted to radio mathematics. The first chapter covers basic electrical concepts while additional chapters are devoted to the direct current theory, basic a.c.

theory, motors and generators, inductance, capacitance, advanced a.c. theory, the vacuum tube, the v.t. oscillator, receiving circuit principles, transmitting circuit principles, antennas, sound conversion, radio aids to navigation, measurements in radio, studio and control equipment, and television and facsimile.

Seven tables in the appendix deal with schematic radio symbols, mantissas of common logarithms, trigonometric functions, db. vs. voltage and power, frequency tolerances, loss constants for attenuation pads, and the international "Q" signal code.

The work is written by and for the practical radio man with the result that all of the material is presented in usable and convenient form. Practical radio technicians, those studying or holding operators' licenses, and practical radio engineers will all find this book valuable in their work.

"RADIO OPERATING QUESTIONS AND ANSWERS" by J. L. Hornung. Published by *McGraw-Hill Book Company*, New York. 574 pages. Price \$5.00. (Tenth Edition.)

This well-known study text, making its appearance for the tenth time, has been completely revised to provide concise information on radio communication—theory, apparatus, circuits, laws and regulations, FM, television, aircraft radio, radar and loran, grouped into the seven elements which go to make up the FCC commercial radio operator licensing examinations.

The new FCC element, Element 7, which became effective on January 3, 1950 is covered by 266 questions dealing with rules and regulations, radio theory operations, and special air navigation problems.

In addition, this new volume includes 46 special problems covering the more complex questions given in the FCC Guide.

Specifically the book provides 1753 answers to the examination elements. In addition much of the data will prove valuable to those in all branches of electricity, radio broadcasting, and marine radio operation. -30-

Members of the Federation of Radio Servicemen's Association group around Robert H. Bishop, vice-president in charge of sales, who accepted the Federation's award on behalf of Sylvania Electric Products Inc. The company was awarded the plaque for outstanding promotion and advertising efforts in stimulating public confidence.





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May, 1950

87



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You Don't Need a FIXED BIAS VS. SELF BIAS SLIDE RULE to Show

Here are some of the advantages of using self bias for audio amplifier power output stages.

By CHARLES P. BOEGLI

Cincinnati Research Company

ROM time to time various pa-pers have appeared stressing the advantages of fixed bias over self bias for audio amplifier power output stages. Generally the writers utilize the data printed in tube handbooks to prove the superiority of the former. For example, in the case of push-pull 2A3's with 300 volts on the plates, a fixed bias of -62 volts and a load resistance of 3000 ohms (plate-to-plate) result in a power output of 15 watts at 2.5 per-cent distortion. With the same plate voltage and a load resistance of 5000 ohms the self bias produced by a 780-ohm cathode resistor (-62.5 volts with zero signal) permits only 10 watts output with 5 per-cent distortion.

Such a comparison is, however, unfair; in the highest-fidelity amplifiers self bias is actually equal to fixed bias in performance and much better from the standpoint of convenience. That self-bias arrangements are more convenient than fixed bias, and consequently to be preferred if all other things are equal, is almost self-evident. Self bias eliminates the need for separate power supply systems with their associated rectifiers and filter networks. Furthermore, since a greater d.c. grid impedance is permissible with self bias, there is no need for expensive transformer coupling or cumbersome cathode-follower drive, which introduces two extra tubes that provide no gain.

With the 2A3, for example, there are two reasons for the poor showing of self bias in the tube manuals. First is the fact that the tubes under the specified conditions are in class AB₁ operation. Under these circumstances the plate current increases considerably as the signal becomes greater, after the signal has exceeded a certain value. This is an advantage with fixed bias1 but quite the opposite for self bias, because here if the bias is correct at zero signal, it will be excessive when a large signal is applied to the grids. The cathode bias resistor thus acts in opposition to the input signal level, resulting in greater distortion for a given output.

The answer to this objection is that in the highest-fidelity amplifiers class AB, output stages simply are not used. The *calculated* freedom from even-harmonic distortion would *actually* be obtained only if perfect coupling existed between the halves of the output transformer primary. This is, of course, never the case. There is no substitute for push-pull Class A output stages when perfect reproduction is demanded. Since with Class A the plate current varies only slightly (because of rectification effects²) with the magnitude of the signal, the bias remains substantially at its predetermined value regardless of the method of supplying it.

In designing a push-pull output stage it must be remembered that the term "plate voltage" on the plate characteristics chart actually refers to the voltage between plate and cathode. When fixed bias is used the 'plate voltage'' is the same as the "B" supply voltage (neglecting the small drop in the output transformer primary) but with self bias, the bias voltage appears between cathode and ground and the "B" supply voltage must equal the "plate voltage" plus the bias voltage. What this means is that instead of furnishing a negative fixed bias by means of rectifiers and filters, we merely add the amount of the bias to the plate supply voltage. We replace all the power supply components of fixed bias with one or two little bias resistors.

For example, if we want to use push-pull 300A's at 400 plate volts and -87 grid volts (plate current = 60 ma. per tube) we can have a 400volt "B+" supply and a -87-volt bias supply, or a 487-volt "B+" supply and no bias supply. In either case we get the same power output and the same distortion—provided just one more precaution is taken with self bias.

This precaution eliminates the last cause of self-bias trouble. If a common cathode resistor is left unbypassed the second-harmonic voltages are fed back in phase to both grids and a secondary type of harmonic distortion appears in the output. Consequently, that resistor (or resistors) is bypassed—the bigger the bypass condenser the better.

The power output from push-pull class A output stages is generally limited by the allowable plate dissipation of the tubes, which must not be exceeded with no signal on the grids. For this reason class A doesn't furnish the power that class AB, does. From 2A3's or 6B4G's only 10 watts maximum is available with 250 plate volts, and an increase in plate voltage much above 250 results in excessive plate dissipation. This merely means that if more power is



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required different tubes must be used, like the 275A (12 watts maximum at 300 plate volts), the 252A (19 watts maximum at 500 plate volts) or the 300A (30 watts maximum at 450 plate volts). Push-pull parallel arrangements can also be used for more power.

In a high-fidelity amplifier with push-pull class A output it thus makes little difference, as far as results are concerned, whether fixed or self bias is used. The only difference is in cost and convenience, where self bias excels. There is no reason for the disrepute that has been attached to self-bias circuits and they deserve every consideration in the design of fine audio amplifiers.

REFERENCES

¹ Terman, F. E.; "Radio Engineer's Hand-book", First Edition, McGraw-Hill Book Company, New York, 1943, Page 383. ² Smith, F. L.; "The Radiotron Designer's Handbook", Third Edition, The Wireless Press, Sydney, 1942, Page 285. –<u>30</u>–

MUTING CIRCUITS By ROBERT HERTZBERG

 ${
m THE}$ amazing ability of an unshielded receiver to pick up signals from extremely distant stations was demonstrated very forcibly recently. While playing with the numerous adjust-ments of a 630TS type television receiver, just to see what happens (and believe me, plenty does!), it was found that the v.t.v.m. gave peculiar readings, some of them not related at all to the circuit functions. In twisting some of the alignment screws, it was discovered that the meter responded very nicely with the test signal generator turned off! After two hours of just plain exploring, mixed with a little swearing, I tuned the mysterious impulse through the sound section of the receiver and nearly fell over when a veddy British voice announced a BBC program originating in London!

The signal was killed by pulling out the oscillator tube, and then the circuits beginning with the i.f. stages be-haved normally. Eventually, I lined the inside of the receiver cabinet with copper screening, grounded thoroughly. This eliminated not only unwanted voice and c.w. signals, but also TV signals being picked up directly by the chassis. These latter signals previously had produced "leading" ghosts. -30-





A

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May, 1950







Sweep Generator (Continued from page 50)

designed to supply only d.c. and filament power. The power transformer is of a rather conventional type except for the several taps on the secondary to supply various 60 cycle sine wave voltages to the sweeping reactor (L_1) . The d.c. is supplied to bias the reactor from the regular "B" supply through R_{30} . The a.c. is supplied through C_{20} via the tap switch S_{4} . Since R_{30} is in series with the d.c. supply, it prevents the filter condensers from shorting the a.c. as supplied to the wide gap reactor. There is considerable a.c. power dissipated in this resistor because the alternating voltage applied to the sweep unit appears across it.

The blanking voltage previously mentioned is obtained from the high voltage secondary via C_{19} and is shaped by the crystal diode $XTAL_2$. Both C_{19} and R_{31} are used to get the blanking phase to correspond to the ends of the trace. Blanking amplitude is not a matter of very great importance because once the oscillator is shut off, it cannot cause any effect on the scope trace until oscillation is again initiated. On the oscillating half cycle, the amplitude of the blanking pulse is carefully controlled by the crystal diode causing a fixed voltage to appear as oscillator bias. A scope applied at pin 6 on the swept oscillator grid will show a half of a square wave during the "on" time, and a half of a sine wave during the "off" time.

Horizontal Sweep

A separate winding is provided for the horizontal voltage for the scope because a 360 degree phase shift is required and a center tap must be incorporated in order to accomplish this. The center tapped winding is not used to supply the oscillators' heaters because every lead into the shield cans



Fig. 8. R.f. output cable and filter.

must be double filtered and it is easier to use an extra winding, thus avoiding one filter.

Phase Control

 R_{37} and R_{38} in combination with C_{25} and C_{26} permit greater than 180 degree control of the horizontal sweep voltage. This assures that for practically all conditions met in the associated equipment, it will be possible to adjust the phase of the scope sweep to match the phase of the radio frequency sweep. This adjustment is easily made by turning off the blanking and adjusting the phase control until the two images are as nearly superimposed as possible. R_{39} and C_{27} serve to remove harmonics which appear in various amounts on the 60 cycle line voltage. C_{27} was chosen to have exceptionally good high frequency characteristics as an aid in minimizing leakage.

The purpose of the phase reversal switch is most easily explained by an example. Let us assume that the Channel 8 response curve of a television set's head end has been set up on the scope with high frequencies on the right. When the above conditions exist let us say the swept oscillator is moving from low to high frequencies covering 175 to 190 mc. Now if the scope were moved to the video detector to view the set's over-all response curve, the high intermediate frequencies would be on the left. This is

Fig. 9. Bottom view of unit showing details of the double shielded construction.





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true because in the usual television set the oscillator is on the high side of the signal. This means then that as the sweep generator's frequency goes up, the intermediate frequency produced goes down, thus turning the curve around. The high frequency side may be returned to the right side by simply flipping the phase reversal switch.

In view of the fact that such reasoning becomes extremely confusing, the following rule of thumb is offered. Turn the sweep generator output frequency higher. The response curve on the scope will then move toward the low frequency side in terms of sweep output frequency.

Spurious Outputs

When working at any other frequencies than those in the range of 165 mc. to 220 mc. a generator such as the one outlined above will have two spurious outputs from the two beating oscillators. The usual practice is to overlook these undesired outputs and depend on the selectivity of the receiver to remove them. Where the most accurate results are desired a filter should be used to attenuate the undesired voltages. Such a filter is shown in the schematic as L_i and C_{18} . Fig. 8 shows the filter with its cover removed inside the loop of the output cable. Some of the undesirable effects which can be caused by spurious outputs of large amplitude are overloading of early stages, excitation of portions of the circuit not under observation, and causing a "hot" output cable.

Conclusion

This article has been confined to discussion of what we believe to be an excellent laboratory-quality sweep generator, but good equipment alone is not sufficient to yield good results. In the field of visual alignment good technique is more important than in almost any other type of electronic measurements. Reliable results can be expected only after a thorough study of available literature on the subject plus many hours of experience.

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NBFM Unit

(Continued from page 63)

changes in plate voltage on the plate of the second section of the 6SL7GT. The v.f.o. in use at W6EDG has the grid circuit of the oscillator on 80 meters with 10 microhenrys inductance and 200 $\mu\mu$ fd. of capacity. The coupling condenser between the v.f.o. and the grid of the 6SL7GT (C_5 , 15 $\mu\mu$ fd.) was chosen by installing a variable trimmer and adjusting to optimum value and then replacing the trimmer with a fixed condenser of the same value. This unit has been used on 10. 11, and 20 meter phone with good reports of narrow-band FM quality.

Note that in the first section of the 6SL7GT a potentiometer is used in the cathode lead to ground. The entire potentiometer is used as a bias resistor and the tap is used for connecting a hand key or bug for frequency shift keying of the transmitter. When part of the bias resistor is shorted out, the plate voltage changes on the second section of the 6SL7GT, due to a change in plate current in the first section of the 6SL7GT. This change in plate voltage changes the amplification and the input capacity will change. The magnitude of the frequency shift will be dependent on the amount of bias change or the amount of the bias resistor shorted out. Note that the bias resistors are bypassed for r.f. only. This FSK system has been used by the writer on 11 meters and works very well. The amount of frequency shift on FSK is variable by means of R_5 from 0 to 3200 cycles, on 11 meters, with the circuit constants shown. Receiving FSK is a matter of having a receiver with good enough selectivity to discriminate between the "mark" or key-down signal and the "space" or key-up signal. The transmitter carrier will be on continuously but the frequency will shift from the mark to the space frequency. It is only necessary



Top view looking down into NRFM and FSK unit bolted in place on front of v.f.o.

for the receiving operator to adjust his receiver so that he receives the mark or key-down signal and the space or key-up signal will then be inaudible. A crystal filter or "Q5-er" will discriminate between the mark and space signal and provide a good signal for copying.

The unit in use in the writer's v.f.o. is built on a small strip salvaged from surplus gear with two conveniently located sockets and tapped holes for mounting. The unit is mounted directly above the tuning condenser of the v.f.o., bolted to the front panel. Filament and plate voltages are taken from the v.f.o. power supply.

The same technique used for communications would also be adaptable for a wireless record player to be used with FM receivers.

It would be desirable to have the oscillator operate on the higher frequency bands, preferably in the FM band of 88-108 mc. As only low power is needed for this type of service, an oscillator such as a 6C4 will be sufficient. With the ordinary crystal pickups it should be practical to feed the pickup directly into the first grid of the 6SL7. The deviation may be adjusted by the cathode potentiometer. -30-

Front view of v.f.o. Controls under dial are deviation control for FSK in center, microphone input on left under dial, and key jack for FSK on right.



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MARS AND NAVY SPONSOR DOUBLE-HEADER PROGRAM MAY 20TH

MARS AND NAVY SPONSOR DOUBLE-HEADER PROGRAM MAY 20TH MARS and the Navy will co-sponsor a double-header program for amateurs on Armed forces Day, May 20, 1950. The two-fold program will consist of receiving competition, sim-ilar to the old Navy-Day Copying Contest, and a QSO and message relay contest which will emphasize the handling of traffic. The Honorable Louis A. Johnson, Secretary of Defense, will issue a special message to he amateurs. This message will be broadcast on 13 military frequencies at 25 w.p.m. as follows: May 20, 1950 at 2100 EST and at 2300 EST. NSS, Washington, D. C., will transmit on 122 kc., 4390 kc., 12,630 kc. and 17,000 kc. WAR, Washington, D. C., will transmit on 3497.5 kc., 6997.5 kc., 12,630 kc. and 17,000 kc. WAR, Washington, D. C., will transmit on stark, and receive and transcribe the message without error may send his tran-scribed copy to Headquarters MARS, Room 5 B 519, The Pentagon, Washington 25, D. C., where it will be checked for accuracy. All who submit a perfect copy will receive a special continue of Merit attesting to their code copying proficiency. The QSO and message relay contest will last 12 hours and will have for its main purpose the demonstration on a national scale of the effectiveness of point-to-point or person-to-person communication by amateur radio as a back-up system for normal communications systems which might be knocked out by flood, fire, storm, ice, or sabotage. The contest will begin at 1700 GCT (1200 EST, 1100 CST, 1000 MST, and 0900 PST). Any radio amateur licensed by the FCC or by the Armed Forces of the United States is eligible to compete in the consist. All amateur bands, either fone or c.w. may be used. Single or multi-operator stotions will be considered separately for purposes of scoring the contest. Contest log forms for additional information on the Armed Forces Day program) may be obtained by writing the contest. All amateur bands, either fone or c.w. may be used. Single or multi-operator stotions will be considered separately for purpo

ING Four Fox Able Charlie, the "MARS" Station of the USAF Special Staff School, Communications-Electronics Division, Gunter Air Force Base, Montgomery, Alabama, has been named the Station of the Month, by Major Rawleigh H. Ralls, Chief MARS, USAF. K4FAC has been "On-the-air" for a year and a half. The station was activated by Lt. Colonel Milburn B. Andrews, Chief, Engineering Section of the school. The original station equipment consisted of a surplus BC-610 transmitter, a "Super-Pro" receiver and a simple dipole antenna. The station was later completely "revamped" and equipped with a new Gordon 10-20 rotary beam antenna, Collins 30-K1 transmitter, National HRO receiver, RME prese-lector, converter, "Panadaptor," and other fine "ham" gear.

Each year two different "K4FAC MARS Clubs" are organized, one from each Senior Air Communications-Electronics Staff Officers' Course. The present club has twenty-two members. Although most of the club members have their own ham stations, most of them will be found on-the-air at

Lt. Colonel Milburn B. Andrews (W4TB), K4FAC station director, hands a message to Sgt. H. B. Solomon, station operator. Sgt. Solomon handles most of the regularly scheduled contacts at K4FAC. Most of the student officers who make up the membership of the MARS Club at Gunter Air Force Base operate in late afternoons or in the evening.



RADIO & TELEVISION NEWS

LOWEST T	UBES an	nd PARTS
$29^{c}_{ea.} \xrightarrow{104}{2256} \xrightarrow{10}{12A} \xrightarrow{112A}{182B} \xrightarrow{112A} \xrightarrow{112A}{182B} \xrightarrow{112A} \xrightarrow$	1A4 1D8GT 1H4G 1A4P 1F4 1H6GT 1B5 1F5G 1J6G 1D5GT 1G4GT 1619 1D7 1G6GT 1626 Sprague 32 35 37 YT-52	El \$20.00 List Value Cornell-Dubilier, Mallory, Aerovox, a, Solar Filter Condensers moving 150V & 450V filters 1B3GT 6G5 150V & 450V filters a 65D7GT 2 58 89 140 cts
$\begin{array}{rrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrrr$	33 3524GT 38 56 34 36 46 57 5 6K7GT 6SQ7GT 12AU7 6 6K8GT 6SR7 12BA6 6L5 6U6GT 12BE6 6 6SA7GT 6V6GT 12H6 1T 6SA7GT 6V6GT 12H6 6SG7GT 6X4 12J5GT 12AG7 6SG7GT 6X4 12J7GT 12AGT 6SG7GT 6X4 12J7GT 12K8GT 6SH7 12A8GT 12K8GT 12F5F 6SH7 12A8GT 12K8GT 12F5F 6SK7GT 12AU6 125F5 12AU6	10 HT-513 6Å7 50A5 12SH7GT 35Z6GT 6AK5 70L7GT 12SR7GT 50B5 6AK5 70L7GT 1629 (eye) 50C5 50Y6 807 25L6GT 50Y6 50X6 77 25X6 77 6BG6G 25BQ6GT 27 78 6BN6 117L7GT 35B5 85 19BG6G 117P7GT
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	6J6 6SS7 7A6 7E6 6K5 6SU7 7A7 7E7 6P5GT 6Q7GT 7B5 7F7 6R7 6T7G 7B6 7G7 6S8GT 6W7G 7B8 7H7 6SF5GT 6Y6G 7C4 7J7 6SF5GT 6Y6G 7C4 7J7 6SN7GT 7A4 7E5 7N7 3LF4 6L8G 7A8 14A4	7Q7 12AV6 12SG7 20 53 7S7 12AX7 12S8GT 35/51 75 7T7 12BA7 12S7GT 35L6GT 84/6Z4 7V7 12BF6 12SK7GT 40 117Z3 7W7 12C8 12SL7 41 VR150 7Y4 12J5 12SN7GT 42 XXL 12A77 12Q7GT 12SQ7GT 43 12AT7 12AT7 12SA7GT 12X7 50L6GT 81
69 c 1AB5 1LC6 1P5GT 1H5GT 1LD5 154 1LA6 1LH4 2V3G Tube prices are for 50 tubes or more- 50L6, 35Z5, 12SK7, {2 10	AA6G 657G 7C5 14A5 6B7 6SU7GT 12A6 14A7 6BF6 6T8 12BF6 14AF7 Less than 50 tubes, 5c per tube extr may be assorted. Individ Miniature_tubes_12AT6, 12BA6, c	14B8 14J7 14Y4 45 83 14E6 14N7 19T8 2050 117Z6GT 14E7 14Q7 25Z5 2051 9001 ra. 14W7 35Y4 50C6 XXB dually boxed—Standard factory guarantee Q0 1U4, 3S4, 1S5, 1R5 14Q
12SQ7, 12SA75 tubes for ?2.13 1R5, 1S5, 1T4, 3V4 Battery Tube Special4 tubes for \$1.49 7JP4\$9.95 10BP417.95 16AP439.50	12BE6, 35W4, 50B5 5 tubes for 1 . 3-Way Portable Tube Kit, 117Z3, 1 . 1U5, 3V4, 1R5, 1T4 all for 1 . 3S4, 1T4, 1S5, 1R5 4 tubes for 1 .	$\frac{4 \text{ tube kit f1.45}{304, 174, 185, 155}{4 \text{ tube kit $1.49}}$ $\frac{304, 174, 185, 155}{5045, 35Y4, 1447, 1486, 1407}$ $\frac{5045, 35Y4, 1447, 1486, 1407}{5 \text{ tubes for $2.95}}$
Best Quality SPEARERS Alnico 5 PM 10 or more Price 5'' - 95c - \$1.05 21/2'', 3'', 4'' - 95c - \$1.05 $\frac{5'}{295} - \frac{51.49}{295} - \frac{51.59}{3.25}$ $\frac{10}{10'} - \frac{295}{4.95} - \frac{3.25}{4.59}$	VERY BEST BRANDS 10 or more Each Price ½ meg, or 1 meg, or 1/10 meg. Each Each with switch—long shaft. 29c 35c switch, long shaft. 29c 35c y meg, long shaft. 10 or 2 meg., long shaft. 29c 35c 1000 ohm 16c 19c 19c 1000 ohm 16c 19c 19c	For 5016, 3516, 50A5, 35A5, 39c ea. For 6V6, 6F6, 305, 304, 354, 3V4, 45c ea. UNIVERSAL OUTPUT TRANSFORMER SPECIAL Up to 12 watts to any speaker (while they last) 98c ea,
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IF TRANSFORMERS Standard Replacement Regular size 455 Kc Midget 455 Kc VIBRATORS	Special on No. 47 Pilot Lights Only- 100 Bulbs\$3.95 Box of 1054 PILOT LIGHTS-100 BULBS \$4.90 Box of 10 bulbs 54	BY-PASS CONDENSERS S 100 Condensers assorted \$5.95 001 Sc 00025 002 Sc 00025 005 Sc 00025 005 Sc 0005 6c
Red Hot Vibrator Special. 4-prong, small size Universal, fits 80% of all jobsea. 89c Jobbers: Write for quantity price. 4 PRONG VIBRATORS—VERY BEST BRANDS Standard Replacement Universal. \$1.29 ea. 4 Prong PHILCO VIBRATOR. \$1.29 ea. RED HOT VIBRATOR SPECIAL! Standard small size, 2%" in height (the popular small size). \$1.29 ea. Stock up while they last, all 4 prong Universal Vibrators.	No. 40 6-8 V.15 Amp No. 44 6-8 V.25 Amp No. 41 2.5 V.50 Amp No. 46 6-8 V.25 Amp No. 51 6-8 V.20 Amp No. 46 6-8 V.25 Amp 6-FT. LINE CORDS Appliance Cord, up to Good Rubber with 1,000 watts.1 rub- 99C Appliance Cord as Un der writers' Approved. UL label 5 for \$1.29 10 for SPECIAL—CONDENSER KITS Stit of 50 BY-PASS CONDENSERS Kit of 50 BY-PASS CONDENSERS \$1.19 Kit of 50 BY-PASS CONDENSERS \$1.95 Kit of 50 BY-PASS CONDENSERS \$1.95	01 500 V. 7c 300 mmt (600 V. 02 8c 100 mmt (600 V. 11 9c 50 mmt (600 V. 12 8c 100 mmt (600 V. 13 9c 50 mmt (600 V. 14 9c 9c 15 9 mmt (600 V. 60 mmt (600 V. 14 9c 9c 15 9 mmt (600 V. 60 mmt (600 V. 16 9c 9c 17 9c 9c 18 9c 9c 19 9c 9c 10 9c 9c 1
OCTAL SOCKETS MOLDED 10 for 49c 7 PIN MINIATURE SOCKETS 10 for 49c 9 Pin Miniature Sockets 10 for 49c Loctal sockets 10 for 49c Standard replacement crystal cartridge. 51.39 NYLON 1J CARTRIDGE 52.59	Kit of 50 MICA CONDENSERS, 1.98 complete 100 resistors—packed in a box, IRC, ett Best values only—½ watt, 1 watt, 2 watt SELENIUM RECTIFIERS Standard 100 mil. Each PUSH-BACK WIRE 100-ft. rolls. 39c ea	VARIABLE CONDENSERS Two gang for superhet or TRF Oscillator Coils for any 5 tube AC-DC Image: Standard Broadcast Band Rated accounts—10 Days All others 20% deposit with order, balance COD Stondard Broadcast Band Content of the order of the ord
FILTER CONDENSERS See Preceding Page	PREMIER RADIO TUBE COMPAN 551 West Randolph St., Chicago 6, Ill Phone: Andover 3-1590 "Your Tube Source Since 1926"	All shipments FOB Chicago. Prompt attention paid to foreign orders. ORDER TODAY. Our parts and tubes are warranted to be 100% replacements for the prototypes in the listings above. Satisfaction Guar- anteed. To speed up delivery, sign your order and your remittance with the same name. Illinois resi- dents add 2% sales tax. CURRENT ADVERTISED PRICES APPLY.



K4FAC at some time or another. Nonamateur students are also encouraged to join the club and regular instruction in code, equipment construction, and station operation is conducted for their benefit.

Master Sergeant J. E. "Mat" Mathis as Chief Operator, and Sergeant H. B. "Sol" Solomon hold down the regular station operating shifts and also double as code instructors for the members of the "MARS" Club. Mat is a handy man with tools and is usually the one who builds "that new gadget for the station." Sol is the traffic man and is also gradually getting the "DX" bug.

Lt. Colonel Andrews has been a ham for twenty-four years with such former calls as W5AQ, W6SU, W5BIT, W5GIK, W2MSY. His present call is W4TB and he will usually be found on 10 or 20 fone chasing that very elusive and exclusive "DX." He holds both commercial radiotelephone and radiotelegraph licenses and shakes a speedy bug when he can be talked into doing a little brass pounding on the circuit.

K4FAC is anxious to schedule any of the other MARS Stations and will handle traffic for any area. The station works with the "Alabama Emergency Net," and the "Montgomery Emergency Net." During the hurricane season last year the station was on 24-hour status for several periods. Future plans for the station include auxiliary power supply for the main station and additional portable and mobile gear to prepare the station for any type of emergency operation.

HAM CONVENTION

THE Detroit Area Amateur Radio Council is sponsoring a ham convention to be held at Detroit's Hotel Statler, May 27, 28, and 29.

This is the first convention held by the Great Lakes Division of the A.R. R.L. since its formation from the old Central Division some four years ago.

Central Division some four years ago. Because of the long Memorial Day weekend, the committee in charge of arrangements has been able to plan some unique and interesting events. The program includes all sorts of technical and organizational talks and meetings. There will be a mammoth "Swap 'N Shop" session to which all who attend are invited to bring their swapping material. An YLRL luncheon, a DX stag breakfast, and an AEC gettogcther have also been planned. Entertainment for XYL's has been provided in addition to events catering to specialized ham interests.

The registration fee for the 3-day meet is \$3.50 with the banquet ducats costing \$4.25—or both for \$7.50. Tickets can be secured by writing direct to W8AHH, Box 903, Detroit 31, Michigan, or can be purchased through radio clubs and radio parts dealers in the Michigan, Ohio, and Kentucky area.

General chairman of the convention is Ty Kirby, W8TDO. Frank Taylor, W8AW, is secretary, and Fred Chevillot, W8SWL, is treasurer. George Goldstone, W8MGQ, is serving as vice-chairman of the affair.

-30-

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HQ-129 Spkr.	11.85	10.00	VFX
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Spkr.	450.00	199.50	(con
SP-200 Series with pwr. sup	op.	100 50	
& Spkr.	250.00	139.50	
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coils for I band	150.00	99.20	DEFCI
MILLEN	_		PRECI
90800 Exciter with tubes	& 10 FO		TACKS
I set of couls	42.50	24.00	PRECI
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HFS VHF Receiver with py	vr.	110.00	
supp.	104.43	110.00	DDDCI
HRO Jr. with B.S. coils	œ	125.00	PRECI
pwr. supp		120.00	per
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(Continued from page 53)

stability, detune C_2 slightly to one side of exact dip. (b) Switch the meter to the doubler stage and tune C_7 for lowest plate current dip. (c) Switch the meter to the buffer stage and tune C_{12} for plate current dip. (d) Switch the meter to the PM section of the final amplifier (shunt resistor R_{16}) and retune each of the preceding exciter stages except the oscillator for maximum upswing of the meter. This will be between 15 and 20 ma. (e) Switch the meter to the PA section of the final amplifier (shunt resistor R_{15}) and tune $\bar{C_{16}}$ for plate current dip. This will be 50 to 60 ma.

5. Connect the antenna. Swing coupling coil L_5 into L_4 to increase coupling (always use the minimum permissible coupling) and tune the antenna circuit, watching for a peak indication of a thermocouple ammeter or lamp connected in the transmission line or feeder cable as C_{18} is adjusted.

6. Plug in the microphone, advance gain control R_{17} , and whistle or hum a sustained constant-volume note into the microphone. Under modulation, the PA plate current will fall to a lower value. The PM plate current at the same time will increase to approximately 4 or 5 times its unmodulated value. The meter or lamp in the transmission line will give an upward indication (an increase of up to 50 per-cent, indicating full modulation). A downward indication of either the lamp or antenna ammeter indicates improper tuning of C_{18} and/or the wrong number of turns in coil L_{5} . The number of turns given for L_5 in Table 1 will be satisfactory for lowimpedance cables such as RG8/U, but the number must be increased for 300-ohm lines or 600-ohm open lines. If downward modulation is obtained and further adjustment of C_{18} does not correct the condition, coil L_3 must be rewound. Modulation quality may be checked aurally by means of a non-oscillating diode (crystal or tube) detector and headphones. Modulation percentage may be checked visually by means of an oscilloscope, using wave patterns obtained with the aid of the internal sweep oscillator of the scope.

7. After tuning adjustments are completed, apply the full plate voltage to the 829B and recheck tuning throughout. Use the PM plate current as an indicator, retuning to give maximum PM plate current.

8. As a final check, the difference between the "in-resonance" and "outof-resonance" unmodulated PA plate current under load should be not less than 5 per-cent and not more than 10 per-cent of the PA operating current.

9. While jacks have been provided for checking PA and PM grid current, these checks are not important to the tuning adjustments and were put in this particular model for laboratory measurements only. They may be omitted if the builder desires.

In the 5 months during which this transmitter has been in use, W6MYR has made approximately 50 contacts for test purposes on 10 meters. Many of these QSO's have been coast to coast and some have been out of the country. Without mentioning beforehand the system of modulation, excellent reports as to carrier strength, talk power, modulation percentage, and speech quality have been received in each instance.

On occasion, W6MYR has operated the supermodulated rig (at 40 watts input) side by side with a standard 40-watt plate-modulated transmitter, switching the antenna from one to the other during a QSO. In these tests, the supermodulated rig has outperformed the standard job, delivering heavier audio at the receiving location even when band conditions were such that the supermodulated carrier seemed weaker momentarily than that of the standard rig. Frequently supermodulation cut through QRM (local at the receiver) when standard plate modulation gave poor account of itself.

In tests at K6AI, 90 watts input to the final amplifier gave excellent supermodulation performance.

Wally Newman, W6FEX, has been operating mobile for the past several months with a duplicate model of this transmitter running about 30 watts input. He reports that excellent results have been obtained and that the transmitter performs very satisfactorily in this type of operation.

While the subject transmitter has been designed specifically for 10-meter operation, the authors can see no reason why the unit should not operate equally well on other bands provided coils are wound for those bands. -30-

TAPE DISTORTION By TED POWELL

HERE are a few more notes on the subject of tape recorder headclogging distortion, originally mentioned by Matthew Mandl in the February issue (page 70). The clogging is also aggravated by

The clogging is also aggravated by the following conditions. I. Finger oil and perspiration coating of tape due to handling; 2. Dust, dirt, and fumes coating tape; 3. Plastic adhesive gumming of tape caused by Scotch tape splices; and 4. Tape coating deterioration and "chalking" caused by coarse tape guides.

It is obvious that it is not enough to simply clean the recording, playback, and erase heads. A clean, soft cloth held gently against the tape while running will clean off the dirt and oil coating. When the Scotch tape splices are reached, the tape pulling mechanism should be stopped and the Scotch tape cleaned to remove plastic adhesive from the outside surface. Next all rough tape guides are polished down with very fine sandpaper and crocus cloth. Care should be exercised so that no abrasive gets into the tape pulling mechanism during this operation. -30-

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Model No. 320-K. An invaluable test unit for service, laboratory and school use. Highly stable Hartley oscillator has range of 150 kc to102 mc with fundamentals to 34 mc. Ideal for quick alignment of FM-AM receivers; provides TV marker frequencies. Also supplies 400-cycle audio output from a Colpitts oscillator. Audio oscillator voltage can be used for testing distortion in audio equipment, bridge measurements. etc. Complete with tubes, portable case, etched panel, all parts, knobs, wire, shielded cable, instructions, diagrams, and operating manual. Size: $10 \times 8 \times 4\frac{34}{2}$ ". For 110-120 volts, 60 cycle AC. Shpg. wt., 10 lbs



VOLT-OHM-MILLIAMMETER

Model No. 511-K. A pocket-size re-liable VOM. 3" meter. Germanium crystal for AC. Ranges: DC v., 0.5-50.250.500.2500; AC and out-put v., 0.10.100-500.1000; DC current, 0.1.10-100 ma, 0.1-10amps; res., 0.500.100,000 ohms, 0.1 meg; decibels, -8 to +55. Complete kit, ready to assemble. Shgp. wt., $3\frac{1}{2}$ lbs. \$14.95 83-153. Only \$14.95

VACUUM TUBE VOLTMETER

Model No. 221-K. An easy-to-build, sensitive VTVM formaking measurements in high-resistance VTVM formaking measurements in high-resistance circuits. 15 different ranges, with DC input resist-ance of 26 megolums. $4\frac{1}{2}$ ' meter is protected against burnout. Includes zero center for align-ment of FM and TV discriminators. Double-tri-ode balanced bridge circuit assures top stability. Ranges: AC-DC volts, 0.5-10-100-500-1000; re-sistance, 0-1000 ohms and 0-1-10-100-1000 meg-ohms. With tubes, all parts, portable case, etched panel, wire, instructions, diagrams and operating manual. Size: 6 x $9\frac{1}{16}$ x 5". For 110-120 volts, 60 cycle AC. Shpg. wt., 10 lbs. **\$73.95** 83-152. Only......\$23.95



NEW 5-INCH OSCILLOSCOPE

Model No. 425-K. A sensational scope with all new features at amazing low price! With Push-Pull de-flection! A precision instrument for TV, FM and AM alignment. Extra sensitive: .05 to .1 volts per inch. Wide range, flat from 5 cps to 500 kc, useful to 2½ mc. Wide range sweep circuit from 15 cps to 75,000 cps. Direct connection to plates of CR tube available at rear of cabinet. Complete with tubes (3-6SN7, 2-6J5, 5BP1, 2-5Y3 rect.), portable case, etched panel, all parts and knobs, wire, instructions, diagrams and operating manual. Size: $834'' \times 131'_8'' \times 17''$ deep. For 110-120 volts, 60 cycle AC. Shpg. wt., 30 lbs. \$39.95 83-155. Only

SWEEP SIGNAL GENERATOR

Model No. 360-K. For visual alignment at TV and FM frequencies between 500 kc and 228 mc. For use with any scope. Sweep width variable from 0-30 mc. Crystal marker with variable amplitude. Provision for injection of external marker. Phasing control. Vernier dial calibrated in frequencies. Complete with tubes (2-6C4, 1-12AU7,6X5GTrect.), all parts, shielded cable, portable case, panel, wire, instructions and op-erating manual. Less crystal. Size: 10 x 8 x 63.". For 110-120 v., 60 cycle AC. Shpg. wt., 12 lbs. 83-159. Only. \$29.95 83-159. Only \$29.95

Complete with Easy-To-Follow Diagrams and Instructions

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this is it-this is the Tuner you designed!



the **CRAFTSMEN RC-10** HIGH FIDELITY FM-AM TUNER

This new tuner was your idea. It is the precisely engineered answer to hundreds of questions ... the solution to scores of problems . . . the outgrowth of countless suggestions we've received from you, Developed from your ideas-and a few of ours-the RC-10 retains every feature of the famous RC-8. And it offers a host of innovations.

- Built-in pre-amplifier compensated for reluctance pickups.
- Automatic Frequency Control entirely eliminates drift, simplifies tuning.
- 5 microvolt sensitivity on both FM and AM.
- 10 kc filter on AM eliminates inter-station squeals.
- Base and treble tone controls for boost, cut, or 20-20,000 cycle flat response.
- SEE . . . the RC-100A ultra-sensitive, custom TV with built-in booster. HEAR ... the RC-2 high fidelity
- amplifier. All units finished in chrome.



New Transceiver (Continued from page 39)

slight alteration was necessary. The assembly was dismantled and two additional leaves, patterned after the originals, were cut from sheet brass. One of these was cut about $\frac{3}{16}''$ shorter than the rest and a drop of solder applied for a contact; the contact on the other new leaf was built up to about 1/8" thickness. The switch was then reassembled using heavy paper for the leaf spacers in addition to the original plastic ones and replacing the plastic bolt insulators with ordinary spaghetti tubing. For the s.p.d.t. section, the new leaf with the large contact was put in place with the contact facing outward. Next, the short leaf with its contact also facing out and on the outside one of the four original leaves was bolted in place. The latter is the common terminal and is in contact with the short leaf while in normal (receive) position. When the switch is pressed, the inside leaf bends outward, making contact with the outside leaf and, at the same time, moving it away from the middle leaf. For the d.p.s.t. section, the three remaining leaves were simply placed one next to the other in such a way as to be normally separated and all in common contact when the butterfly knob is turned. Details of the switch before and after alteration are shown in Fig. 3.

The r.f. section came next. The submidget, two-plate variable condenser was mounted on two aluminum brackets which were held in place by base mounting bolts and the coil was soldered across it. The plate pin of the 958-A was soldered directly to a stator terminal and the plate r.f. choke was connected to the center of the coil. The grid resistor, condenser, and choke were soldered to the grid pin and all other leads kept short. A piece of %" polystyrene rod, one end of which was drilled to fit the quarterwave antenna snugly, was used for the radiator socket. The threaded metal part of a tip jack was heated and

Fig. 3. Conversion details on the handset switch. The unit at the left is the original switch while the one at right shows changes made in conversion process.





Fig. 4. The r.f. portion of the transceiver. The r.f. grid choke is in the foreground. The grid-leak resistor and feedback condenser can be seen at the left and above the grid choke. The 958-A plate pin (other side of tube) is soldered directly to the stator terminal of the variable condenser. The base was drilled and tapped to accommodate 6-32 machine screws which secure the antenna socket clamp.

turned into the other end after drilling the plastic to accommodate it. A hairpin turn of No. 14 copper wire, connected between the jack and a ground held under a mounting bolt, provided ample pickup. The antenna was cut from the small end of a steel whip but could have been any conductor of equal length and reasonable rigidity. The condenser shaft extension is $\frac{1}{4}$ " polystyrene rod.

A four foot length of six conductor cable was used to connect the hand unit to the battery pack. Four of the six conductors were fastened to lugs in the microphone chamber and the cable terminated at a shielded octal connector which plugs into the battery box. This box, housing batteries and audio components, was constructed of light gauge aluminum and measures $4'' \ge 7\frac{1}{2}'' \ge 1\frac{3}{8}''$. Its size could have been reduced had smaller batteries and transformer been used. The 1S4 provides ample audio for modulation and reception; in fact the gain had to be lessened to prevent excessive frequency modulation of the carrier and to eliminate feedback from headphone to microphone.

Battery current drain is quite low; approximately 15 ma. for the "B" battery and 200 ma. for the "A" battery. The quarter-watt input to the 958-A does surprisingly well. Considerable distances have been covered under most adverse conditions and arm fatigue is a thing of the past. The enjoyment and convenience in operating this little rig has been ample reward for the slight expense and short time required for its construction.

Some adjustment of antenna length may be necessary for optimum results. However, in most cases the rod should be about 19 inches long. -30MONEY BACK GUARANTEE—We believe units offered for sale by mail order silvand be sold only on and Back-If-Not-Satisfied" basis. We carefully check on the design, calibration and value of all items advertised by us and unhesitatingly offer all merchandise subject to a return for credit or refund. You, the customer, are the sole judge as to value of the item or items you have purchased.

THE NEW MODEL TV-10 THE NEW MODEL 247 SPECIFICATIONS:



Model 247 comes complete with new speed-read chart. Comes housed in handsome hand-rubbed oak cabinet sloped for bench use. A slip-on portable \$2990 ningod cover is indicated for outside use. Size: 103/"x83/"x 53/4". ONLY.....

The New Model 200 AM and FM

• TUBES USED: -12AU7-One section is used as oscillator and the second is modulated cathode follower. T-2 is used as modulator. 6C4 is used

The Model 200 operates on 110 Votts A.C. Comes complete with output cable and operating in-structions.

as rectifier

THE NEW MODEL 670

Check octals, loctals, bantam jr., peanuts, television minia-tures, magie eye, hearing aids, thyratrons, the new type H.F. miniatures, etc.

Features: ★ Newly designed element se-

★ Newly designed element se-lector switch reduces the pos-sibility of obsolescence to an absolute minimum. ★ When checking Diode, Tri-ode and Pentode sections of multi-purpose tubes, sections can be tested individually. A special isolating circuit al-lows each section to be tested as if it were in a separate envelope.

as if it were in a separate envelope. ★ The Model 247 provides a supersensitive method of checking for shorts and leak-ages up to 5 Megohms be-tween any and all of the ter-minals. ★ One of the most important-improvements, we believe, is the fact that the 4-position fast-action snap switches are all numbered in exact accord-ance with the standard R.M.A. numbering system. Thus, if the element terminating in pin No. 7 of a tube is under test. button No. 7 is used for that test.



The Model TV-10 operates on 105-130 Volt 60 cycles A.C. Comes housed in a beautiful hand-rubbed oak cabinet complete with portable \$3950 NET

Tests all tubes including 4. 5, 6, 7, Octal, Lock-in, Peanut Bantam, Hearing-aid, Thyra-tron, Miniatures, Sub-Mini-atures, Novals, etc. Will also test Pilot Lights.

Tests by the well-established emission method for tube qual-ity, directly read on the scale of the meter.

Tests for "shorts" and "leak-ages" up to 5 Megohms.

ages" up to 5 Megohms. Uses the new self-cleaning Lever Action Switches for in-dividual element casting. Be-cause all elements are num-bered according to pin-number in the RMA base numbering identify which element is un-der test. Tubes having tapped filaments and tubes with fila-ments terminating in more than one pin are truly tested of the pins may be placed in the neutral position when nec-essary.

essary. The Model TV-10 does not use any combination type sockets. Instead individual sockets are used for each type of tube, Thus it is impossible to dam-age a tube by inserting it in the wrong socket.

Free-moving built-in roll chart provides complete data for all tubes.

Newly designed Line Voltage Control compensates for vari-ation of any line voltage be-tween 105 Volts and 130 tween Volts.

Enables alignment

of television I. F. and FRONT ENDS without the use of an oscilloscope. SPECIFICATIONS Frequency Range: 4 Bands—No switching

18-- 32 Mc. 35- 65 Mc. 54- 98 Mc. 150-250 Mc.

SIGNAL GENERATOR +

R.F. FREQUENCY RANG-ES: 100 Kilocycles to 150 Megacycles. MOD UL ATING FRE-QUENCY: 400 Cycles, Also available separately. ATTENUATION: The con-stant impedance attenua-tor is isolated from the os-cillating circuit by the buffer tube. Output im-pedance of this model is only 100 ohms. This low impedance reduces losses in the output cable. OSCILLATORY CIRCUIT:

.

the output cable. **OSCILLATORY CIRCUIT:** Hartley oscillator with cathode follow-er buffer tube. Fre-quency stability is assured by modulat-ing the buffer tube.

assured by modulat-ing the buffer tube. ACCURACY: Use of high-Q permeability tuned coils adjust-ed against 1/10 of 1/% stand-ards assures an accuracy of 1% on all ranges from 100 Kilo-cycles to 10 Megacycles and an accuracy of 2% on the higher frequencies.

SUPER

METER

DECIBEL MEASUREMENTS, D.C. VOLTS: 0 to 7.5/15/15/15/150/750/ 1500/7500 A.C. VOLTS: 0 to 15/30/ 150/300/1500/3000 Volts. 0 U T P U T VOLTS: 0 to 15/30/150/300/150/3000. D.C. CURRENT: 0 to 1.5/15/150 ma.; 0 to 1.5 Amps. RESISTANCE: 0 to 500/ 100,000 ohms, 0 to 10 Megohms. CA-PACITY: 001 to 2 Mfd., 1 to 4 Mfd. (Quality test for electrolytics.) REACT-ANCE: TO0 to 27.000 Ohms; 13.000 Ohms to 3 Megohms. INDUCTANCE: 1.75 to 70 Henries; 35 to 8.000 Henries.

DECIBELS: -10 to +18. +10 to +38, +30 to +58.

Combination VOLT-OHM-MILLI-AMMETER plus CAPACITY RE-ACTANCE, INDUCTANCE and DECIBEL MEASUREMENTS.



10

Model TV-30 comes com \$29% plete with shielded co-axial lead and all operating instructions.

Tubes Used: 6C4 as Cathode follower and modu-lated buffer. 6C4 as R.F. Oscillator. 6SN7 as Audio Oscillator and pow-er rectifier.

Audio Modulating Frequency: 400 cy-cles (Sine Wave) Attenuator: 4 posi-tion, ladder type with constant im-pedance control for fine adjustment.





The New Model TV-30 TELEVISION SIGNAL GENERATOR



Compact, measures 3¹/₈" x 5¹/₈" x 2¹/₄". Uses latest design 2% accurate 1 Mil. D'Arsonval type meter. Same zero ad-justment holds for both resistance ranges. Justment holds for both resistance ranges. It is not necessary to readjust when switching from one resistance range to another. This is an important time-sav-ing feature never before included in a V.O.M. in this price range. Housed in round-cornered, molded case. Beautiful black etched panel. Depressed letters filled with permanent white, insures long life even with constant use. Specifications: 6 A.C. VOLTAGE RANGES

Inte even with constant use. Specifications: 6 A.C. VOLTAGE RANGES: 0-15/30/150/300/1500/3000 volts. 6 D.C. VOLTAGE RANGES: 0-7½/15/75/ 150/750/1500 volts. 4 D.C. CURRENT RANGES: 0-1½/15/150 Ma.0-1½ Amps. 2 RESISTANCE RANGES: 0-500 ohms, 0-1 Megohm

The Model 770 comes com-plete with self-contained bat-teries, test leads and all op-erating instructions. \$**13**90

The model 670 comes housed in a rug-ged, crackle-finished steel cabinet complete with test leads and operating in-structions. Size 51/2" x Megohm structions. 71/2" x 3". 20 % DEPOSIT REQUIRED ON ALL C. O. D. ORDERS

DEPT. RN-5, 98 PARK PLACE, GENERAL ELECTRONIC DISTRIBUTING CO. DEPT. RN-5, 9 New York 7, N. Y.

May, 1950



HENRY HAS THE <u>NEW</u> hallcrafters MODEL 5X-71 NOW!



This new type of receiver—the first of its kind on the market—has extra sensitivity, selectivity, and definitely superior image rejection. Continuous AM reception from 538 kc to 35 Mc, and 46 to 56 Mc. One RF, 2 conversion, and 3 IF stages. 105-125 volts AC. 11 tubes plus voltage regulator and rectifier. Only \$179.50.

MEDIUM PRICED HALLICRAFTERS MODEL S-40A



540 kc. to 43 Mc. Temperature compensated. One RF, 2 IF, 3-watt output, 4 bands. 115 V. AC. 8 tubes plus rectifier. Internal speaker. Only \$79.95. Other popular Hallicrafters models: \$38-A, only \$39.95; \$-72, only \$89.95; \$X-43, only \$159.50; \$X-62, only \$269.50.

I have a complete stock of Hallicrafters receivers and transmitters. I'll make you the best deal on a trade-in for your communications receiver. I give you prompt delivery, and 90-day FREE service. Nobody can beat Bob Henry on a trade-in, and I offer you the world's lowest credit terms. Write, wire, phone, or visit either store today for the best deal.



The Genescope (Continued from page 62)

The ability of the Genescope to provide a zero or reference line is most appreciated when the FM sound discriminator is being adjusted. Consider the typical television sound i.f. system shown in Fig. 6. It consists of three i.f. stages and a discriminator. The 3rd i.f. tube serves as a limiter to reduce the effects of amplitude modulation. The output cable of the sweep oscillator is connected to the grid of the 3rd i.f. amplifier tube. The vertical input cable of the oscilloscope is connected between point A and ground. The sweep generator is set to sweep for approximately .5 megacycle above and below the sound i.f. frequency. The vertical gain control on the oscilloscope is now advanced until a response pattern approximately 1 inch high is obtained. The pattern will consist of two separate S-shaped response curves. Rotate the phasing control until the two curves blend as much into each other as possible. The result is shown in Fig. 7A. Now adjust the blanking control until a single curve appears with a base line through it. See Fig. 7B.

Adjust the primary of the discriminator transformer for maximum amplitude and linearity of the S-curve. Reduce the signal generator output as the height of the curve increases. The secondary of the discriminator transformer controls the centering of the S-curve and it is adjusted until there is as much of the curve above the base line as below. Feed in a 21.25 mc. marker signal. If the discriminator is properly adjusted, the pip will appear along the linear portion of the "S" curve. As a further test of the secondary alignment, set the AM signal generator to the sound i.f. value (21.25 mc.) and modulate it with the 400 cycle note. If the discriminator is not in perfect alignment, a 400 cycle modulation pattern should appear along the base line of the S-curve. See Fig. 7C. Adjust the discriminator secondary to the point where the 400 cycle disappears, then reappears if the adjustment is continued in the same direction. When the signal is kept as low as possible, this check is a very sensitive one.

A common difficulty encountered by the service technician when attempting to align the r.f. end of a television receiver is the achievement of a prop-

Fig. 9. Impedance matching output cable.



-THE BEST IN ELECTRONIC SURPLUS -

12 V. DC HEAVY BEAM MOTOR

Rated at 1/12 horsepower at 1725 rpm, but actually with built-on reducing gears develops many, many times more power. Worm gear at end of reducer is 43 rpm. As is, can be used for hoisting, electric platforms, portable-electric trucks, conveyors, etc. Operates from 12 volts DC (two 6 volt storage batteries). Additional

7 inch heavy brass worm gear available to reduce rotation to 21/2 rpm, for antenna rotator NEW, unused surplus. PRICE, EACH MOTOR. 7-inch Brass Worm Gear, for 2½ rpm. \$1.50

Hi-QUALITY COMPASS AND ADJ. TRIPOD

For orienting beam antennas, boat owners, hikers and many other useful applications where a real hi-quality compass is required. Compass is deluxe with ball and socket mount adapter, built-in levels, a zero set, a screw for locking needle when not in use, two vertical sights—folding type, 4 inch pointer on 5 inch compass, 0-360 azimuth calibration (coun-ter-clock-wise). Tripod is execedingly well made, telescoping from 20 inches to 47 inches, and can be adapted for camera work. PRICE, Complete, as above, NEW, unused

VARIABLE AUDIO OSCILLATOR



For service test bench, code practice (by keying), tone modulation of ham or commodulation of han or com-modulation of han

\$5.95 PRICE, EACH





C-VOLT VIBRATOR POWER SUPPLY, PE-157 Designed for the 'Horsy-tark' or BC-745 Transmit-tarke' or BL-745 Transmit-tarke' or BL-745 Transmit-tarke' or BL-745 Transmit-tarke' or BL-745 Transmit-or BC-745 Transmit-tarke' or BL-745 Transmit-beliver approx. BL-745 Transmit-tarket approx. BL-745 Transmit-ta

EACH. \$4.95 DECK ENTRANCE INSULATORS, bowl and flange type, 8%" dia, with heavy galvanized metal flange and bell. Top bell 6%" dia, 11%" brass feed-thru rod, Very high voltage insulation. Individually packed in cartons, all NEW. 12 FOR.

11P., generator output 2007. S60.00 NEW GENERAL ELECTRIC AMPLIDYNE MODEL SAM78AB47 MOT OR GENERATOR SET. Motor 3HP 440V, 3 phase. Output 250V. DC at 3 amps. anp 60V. DC at 12.5 amps. Excellent Condition... \$55.00





ARRAY For the ham, experimenter, or com-mercial application where directive transmission and reception are de-sired. This array is designed for 20 to 40 mc coverage. Assembly is simple, and is easily changed to horizontal. Center supporting mast is vertical dipole, contains within an f transmission line, and at top a movable, and design is for both transmission and recep-tion. Sturdy, and ideally adaptable for rotating beam array when used with motor-driven mount below. With matruetion and diagram sheet. Shpg. wt. 75 lbs. PRICE COMPLETE ARRAY



WAVEMETERExampleExa

EACH. \$75.00 **RADAR TRAINING SET-MARK V**

RADAR TRAINING SET—MARK V For Student, Schools, Labs, or actual radar application. Operates in the 580 to 765 me region, designed especially to illustrate how radar eqpt. functions, and permits mak-ing numerous experiments to put over radar funda-mentals. Uses above described receiver; separate trans-mitter using 8025 triode with 1.5 waths at antenna, with 400 KC internal modulator using two 811 tubes; Ex-ternal modulator, generating audio frequencies of 16,000, 4,000, and 1,000 cycles, and RF at 750, 350, and 175 KC, with selection switch for modulation frequency and wave form control, using 3 tubes, 807, 6N7, and 6J5; set of Antenna Dipole rods. Supplies with full instruc-tion sheets, diagrams, calibration curves, tubes, NEW unused surplus. used surplus.

Technical Manual (Complete Instruction Book) for the BC-312 & BC-342 Receiver. PRICE, Postpaid.....\$1.50

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"SNOOPERSCOPE." TUBE Infra-Red Image Converter Tube (British) to make "Snooperscopes," "Sniperscopes," and other devices that see in the dark. Has many useful industrial applications. Operates with invisible infra-red rays, without scanning or amplinders. Supplied with teen-nical data and diagrams. Every tube guaranteed! WE WILL NOT BE UNDERSOLD. Sto

WE WILL NOT BE UNDERSOLD. 55.00 6 for 552.00 BAUSCH & LOMB Front-End Lens Assembly, for best images. F2.1.3.5 in E.F. EACH 510.00 MOUNTED LENS UNIT, also for front-end, results as good as B & L unit. Speed F1.9 f.1.91.44 mm, outside dia.at one end 60 mm, length of mount 64 mm. PRICE, EACH. 57.00

All Above Material Subject To Prior Sale. 25% Minimum Deposit With All C.O.D.'s. All prices F. O.B., N.Y.C.



20-40 MC FIELD STRENGTH AND WAVEMETER

Uses a 0-100 Micro-ammeter with a 1S4 Pentode, to receive signals in the 20-40 mc range. 1.5 volts battery required. Tuning dial has dial lock, for fixing position, and telescopic

position, and telescopic Antenna permits ad-justing for strong or weak signals. Calibra-tion inust be self-per-formed. With instruc-tion sheet and diagram NEW, unused eqpt. Dim: 0.1²/3'4/4"x51/2" (less antenna). PRICE, Each

\$12.95



VEHICULAR FM TRANSMITTER-**RECEIVERS, BC-620 & BC-659**



Compact, almost portable, NBFM transmitter-receiv-ers designed for vehicle in-stallation. Provides instan-taneous selection of either one of two channels, preset for both Xmttr and revr., crystal controlled. Uses 13 selected Loktal tubes in a highly engineered and effi-cient unit to give a 10 to 15 mile range with a minimum of power consumption. On-

WESTERN ELECTRIC 500 WATT "HLAS" ANNOUNCING SYSTEMS

Gigantic Sound Power amplification for giant political rallys, airports, doeks, stadiums, etc. 3-unit eqpt., with 120 db gain; automatic volume control and compression cir-cuits; Speech, Phono, and wire or tape recording inputs; 20 KC erasing oscillator; 12-Speaker bank for handling full output. All in solid, splashproof cases. Operation from 115/3/60 AC.

WRITE FOR FULL DESCRIPTION SHEETS AND PRICES.

PORTABLE 250 WATT W.E. BEACHMASTER SOUND EQPT.

Super-Power 250 Watt Amplifier and Loudspeaker System Super-Fower 250 Watt Ampinier and Loudspeaker System for stadiums, ball parks, airports, etc. Complete pre-amp and power amplifier in water-tight ease with internal blower. 9-Speaker horn assembly in similar case, making complete system portable. Operates from 110/1/60 AC. Complete with cables, tubes, dynamic microphone. WRITE FOR DESCRIPTIVE SHEET AND PRICES.

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IN ELECTRONIC HISTORY! We invite you to come in and browse around, and see our terrific variety of electronic eqpt., all priced way down for quick sale. Intricate, expensive eqpt. costing hundreds of times our special low prices, just a few of each so that it doesn't warrant advertising or cata-loging. You will find, at ridiculously low prices, Tuning Fork Oscillators, Time-Delay Relays, Power Trans-formers of all sizes for transmitters and receivers, an amazing variety of hi-quality audio components, variable condensers of all types and sizes, insulators for every purpose, test and measuremental instru-ments; Commerical transmitters, from ½ watt to 300 watts which will make fine ham rigs, mine detectors, sonobuoy transmitters, radar, underwater sound de-vices, power supplies, metering panels, amplifiers, and hundreds of other devices and parts. PLEASE DO NOT WRITE FOR CATALOGUES OR PRICES. All of this material is priced for immediate take away; no shipping or mail order. Communication Companies and

Communication Companies and Governmental Agencies!

Do we have you on our mailing list? We have, for im-mediate delivery, some large hi-powered commercial transmitters, complete ship rader sets, 10 KVA Gas Engine Generator Sets that are NEW and Export Packed, and many other hard-to-get surplus items. Write us your requirements.





keep with it!

Be sure you get that famous ALTEC 21B quality from over-active announcers and recording artists by using the new 155A Chestplate. Because of the microphone's perfect positioning, you may readily obtain the sound separation you desire from vocalists working with an orchestra... without false bass! With this new adaptation of the 21B, sports announcers can always override even the noisiest crowds, for the 21B does not limit at audio peaks. Its smooth frequency response permits use in high-level sound fields which would ordinarily cause acoustic feedback.

The 155A Chestplate is compact, lightweight and contains a matching unit which permits its use up to 400 feet away from associated equipment.



Send for brochure giving technical information on ALTEC 21B Miniature Microphone adaptations.



1161 N. VINE ST., HOLLYWOOD 38, CALIF. 161 SIXTH AVE., NEW YORK 13, NEW YORK

TERMINATION	CONNECTIONS		
300 ohms	Jumper 1-2	Jumper 5-6	
150 ohms	Jumper 1-2 Jumper 5-6	50 ohms 2-3 75 ohms 4-5	
75 ohms	Jumper 1-2-3	Jumper 4-5-6	
50 ohms	Jumper 1-2 Jumper 4-5	30 ohms 2-3 50 ohms 1-5	
Open Termination	Jumper 1-2-3	Jumper 4-5	
Series Condenser .02 μ fd. 400 v.	Jumper 2-3	Jumper 4-5	

Table 1. Terminal box connections for the impedance matching output cable.

er match between the signal generator and the receiver input terminals. Any appreciable mismatch will act to alter the shape of the response curve, causing adjustments to be made which will not provide the proper response when the signal generator is removed and the regular set antenna is connected. In the Genescope, matching of impedances is specifically provided for by employing the output cable termination shown in Fig. 9. A variable matching network and a series of terminals are available at the end of the generator cable and by connecting the terminals as indicated in Table 1, impedance values of 50, 75, 150, and 300 ohms can be obtained. Information is also provided in the instruction booklet accompanying the instrument for matching to other impedances.

While not specifically outlined above, alignment of FM receivers would proceed in a similar manner.

The stress, thus far, has been directed toward alignment but the Genescope, through its separately contained oscilloscope, is equally valuable for servicing. Waveform checking and testing—which is a common practice for the video amplifiers and the vertical and horizontal sweep systems—can be accomplished by feeding whatever signal is present to the vertical amplifiers of the scope. Where signal tracing through the i.f. and r.f. circuits is required, recourse can be made to the high-frequency probe. This probe is essentially a high frequency detector and may be used to pick up the signal at any point in the receiver where high-frequencies exist. This applies with equal force to video signals received from a station or an r.f. signal obtained from the AM generator section of the Genescope.

To trace a signal through the sound channel of a TV receiver, connect the output cable of the AM generator section to the antenna terminals of the receiver and set the AM generator to the sound r.f. carrier frequency of the channel being tested. Modulate this signal with the 400 cycle note.

Attach the high frequency probe cable to the vertical input jack and touch the probe to various points along the i.f. system. Starting with the converter grid the signal may be traced through the i.f. system to the discriminator.

The same test may be applied to the video i.f. system by tuning the AM generator to the video carrier frequency and proceeding from the converter tube through the picture i.f. stages to the video detector. Further tracing of the signal beyond the video detector can be done with a pair of conventional leads.

Any single stage may be checked by connecting the h.f. probe across the output of the stage under test and injecting a signal of proper frequency to the input. When a received video (or audio) signal itself is being traced through the various stages, the AM generator is turned off and only the scope section is utilized. -30-

The Air Materiel Command at Wright-Patterson Air Force Base in Dayton has recently announced the development of a "Tom Thumb" synthetic radio tube, about the size of a match head, which may prove to be the Air Forces' answer to the problem of how to reduce the weight and size of its airborne electronic equipment. Named the "fieldistor", the sub-miniature tube is now in its early stages of development. The new tube is 1/90th the size of the standard radio tube shown at the left while a presentday electronic tube towers over it at the right. Ten of these units could be carried in an ordinary thimble. Another advantage claimed for the new "tube" is that it uses so little current that most batteries can be eliminated, thus reducing equipment size.





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Spot Radio News (Continued from page 18)

private demonstration in the Hotel Statler, before a group of newspaper men and members of Congress, including the chairman of the Senate Interstate and Foreign Commerce Committee, Senator Edwin C. Johnson. The demonstration, which featured a new approach to the problem, involving a vertical shift rather than the horizontal method used previously, was applauded by the Senator, who declared that he was well satisfied with the system now.

THE HEARINGS disclosed that many of the questions which were puzzling the Commission, particularly those raised by Coy during his Oklahoma address, might have effective answers very soon. Dr. E. W. Engstrom provided the clue that solutions were perhaps not too far away, by declaring that *RCA* had reached the stage where it might be possible to begin probing color standards. Such a step could now be considered, he stated, because of the high-level sampling synchronization and color phasing techniques recently developed and particularly because of the new direct-view, tricolor tube, which would be demonstrated in Washington in the very near future.

Testimony revealed that the directview, single tube shows . . . "an attractive picture with good resolution and good color."

Engstrom reported that ... "Experience in black and white television has shown that in order to realize consistent performance at the limit of system capabilities, it is necessary for the terminal units—pickup tubes and picture tubes—to have resolution capabilities in excess of those of the system itself. Since this requirement seems equally applicable to a color system, our work on the tri-color tube has as a goal, the attainment of such extra resolution capability. We know of no reason why this goal cannot be attained in the near future. . . . Our investigation of receiver circuits for use in connection with a tri-color tube have, of course, been carried out in parallel with the tube development. We feel that this circuit development is straightforward and will provide for further color-receiver simplification. For example, only one yoke and deflection system are used, just as in ordinary black and white receivers. In turn, the elimination of two of the three deflection systems required in the current three picture-tube assembly also reduces power-supply needs. . . . With a suitable tri-color tube incorporated in the simplified receiver, we may produce in the not-too-distant future a fully-electronic color television receiver, with but ten to fifteen tubes more than for a comparable black and white receiver."

Questioning by Commissioners Jones and Hennock, as to the use of the tricolor tube for other systems, brought the striking reply that the tube could be used with the *CBS* system, even though 405 lines were involved, but that performance might not be too satisfactory because of the reduced resolution afforded by the 405-line picture.

When queried about the size of the picture that the tri-color tube would provide, Engstrom declared that a sixteen-inch envelope was being used, with the viewing area masked down to about fourteen inches. The larger envelope had been selected because it simplified construction of the elements required for three-color presentation. Dr. Peter C. Goldmark, the field-

The "electronic umpire" developed by the General Electric Company and used by the Brooklyn Dodgers during spring training. When a pitched ball passes through the strike zone, the shaded area, its shadow is seen by three electric eyes which look at the sky at the angles shown. If the pitch is a strike the ball is seen by the eyes in 1-2-3 order and electric impulses are created which light a "strike" indicating lamp. An inside or outside ball is not seen at all and a high or low pitch results in the ball being spotted in an improper sequence, and the lamp does not light, thereby indicating a "ball" was pitched. The strikes, as well as the speed of the ball, are registered on the recording machine shown in photo at the left.


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1B26	3DP1-S2A 2.79	300 B	836A 1.97	CK502AX 2.25	1A6	6AD7G 1.09	6T7G	1417
1B29	3DZ1A	304 TH	837 1.19	CK503AX 1.95	1A7GT	6AF6G	6U5G	14N7
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1021 3.75	3HP7 3.95	327A/5C37 2.75	851 12.95	CK512AX 2.25 CK571AX 8.45	1B5/25S	6AJ5	6V6GT	24A
1N21	4-65A	331A	860	CK1005	1050	6AK6	6W7G	25255
1N21B 1.39	4-250A	350B 1.98	861 9.95	CK1005	1C7G	6AL5	6X4	25Z6GT 43
1N22	4AP10 1.98	368A/S 4.98	865	E1148	1D5GP	6A06	6X5G1 47	26
1N23	4B22 9.98 4B24 1.98	3/1B	866A 1.05	EF50	1D8GT	6AR5	6C7G	28D7
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1N27	4B26 4.49	394A 3.69	872A 1.12	F128A	1C4GT	6AV6	7A6	32
1P23	4B32	446A	876	F606	1G6GT	6B4G	7A7	32L7GT
1P24	4C35	446B 1.79	878	F862A	1H4G	6B7	7B4	33
1821 3.95	4022 9.95	4501 H	884 1.19	FG17	1H5GT	6B8G	7B5	35/51
2AP1 3.89	4E27/257B 12.45	527 5.95	902	FG32 4.95	116G	6BF6 52	7B6	35A5
2C21/RK33	5AP1 2.95	559	905 2.49	FG33 7.95	1L4	6BF6	7C4	3505
2C26A	5BP1 2.29	631P1 3.75	918 4.95	FG91A 3 29	1LA4	6BC6G 1.47	705	35L6
2C34/RK34	5BP4 2.39	700A 19.95	919 1.95	FG105A 8.95	1LB4	6BJ6	767	35W4
2039	5CP7 2 95	700B 19.95 700C 19.95	923	FG172	1LC5	6C4	7E6	35Z3
2C43	5D21	700D	930	GL146 9.95	1LD5	605	7E7	3575
2044	5FP7	701A	931A 2.98	GL434A 2.69	1LE3	6C8G	7H7	36
2C51 5.95	5JP1	703A 1.89	955	GL4/1A	1L5	6D6	7K7	37
2D21	5JP2 8.95	704A 1.49	956	GL697	1LN5	6E5	7N7	39/44
2E22 1.19	5J23	705A	958	HF100	1N5GT	6F5	707	41
2E26	5J29	706CY 18 75	959	HF200. 14.95	1Q5GT	6F6GT	7K7	42
2J21A	5LP1 12.95	706FY 47.50	991/NE16	HF300	1R4	6F7	7W7	45
2122 7.95	5MP1 9.95	707B 14.95	1614 1.35	HY115	185	6666	7X7	45Z3
2126 6.95	5NP1 4.98	708A	1616	HY615	1\$5	6H6	7Z4	46
2J30 49.50	6F4	714AY 3.59	1624	KU610	114	615	12A	47
2132 12.95	614	715A 5.49	1625	KC4/ML10037.50	104	6J5GT	12A7	50 1.39
2133 18.75	9GP7	715C 19.95	1629 19	ML101	1V	616	12A8GT	50A5
2134	9JP1 6.95	717A	1630	REL21	2A4G 1.07	6J7GT	12AT6 44	50B5
2J37	9LP7 1.98	723A/B 12.95	1632 69	RK59	2A5	6K5GT	12AT7	50Y6
2J38	10BP4 19.69	724A/B 2.95	1633	RK65	2A7	6K7	12AU6	56
2140	10Y	725A 6.45	1634	RK72	2V3G	6K8	12AV6	57
2346	12GP7 12.75	726B	1636 1.98	RX21 2 39	2X2A	616 105	12BA6	58
2149 39.50	12HP7	7260	1638	RX120 8.95	3A4	6L.6G	12C8	70L7
2150	15E 1.19	750TL	1654 2.45	V70D	3A5	6L6GA	12F5GT	71A
2J54B	15R	800 1.49	1665. 1.10	VR53	3B7/1291 29	6L7G	12J5GT	75
2161	1978	801A	1851	VR78	3D6/1299	6N7	12J7GT	77
2162	23D4	803	2050 1.19	VT158	304	6R7	12K/G1	78
2K28	28D/	804	2051	VUII1	3Q5GT	6S7G	1207	81 1.25
2K29	35TG 5.95	807 1.09	8011	WL468	3V4	6SA7 44	12SA/	82
3API 4.59 3B22 1.98	45 SP	808	8012 1.39	WL531 4.75	5R4GY 1.09	6SC7	12SF5	83V
3B24 1.49	75TL 5.95	810 7.95	8014	WL532 1.89 WL616 37.50	5046 49	6SD7GT	12SF7	84/624
3B25 4.87	100R	811 1.98	8020	WL619 18.95	5V4G	6SF7	12SH7	85
3820 1.49	100TH 9.95	812 2.45	8025 3.69	WL677	5¥4G	65G7	12SJ7	117L7/M7 1.19
3B28 7.95	211 25	813 6.86	9002	WL681/68622.50	5Y3GT	6SJ7	12SL7	11/N/
3BP1 2.39	217C 9.95	814 1.98	9002	WL/IUA	5Y4G	6SK7GT	12SN7	117Z3
		time the first rise 1		042 1.23	523	051/01	12SQ/ 49	117Z6
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May, 1950



problem with the new VEE-D-X antenna switch. Just turn the knob, and you can change over instantly from one antenna to another. Eliminates entirely the fuss and bother of changing transmission lines every time a different antenna is used. Very useful for T V Dealers when demonstrating more than one receiver from a single antenna.

Here are the features that make the VEE-D-X Antenna Switch the finest available:

- Specially designed switch prevents leakage.
- Furnished in attractive ivory plastic case with satin finished aluminum face.
- Terminal strip accommodates three separate lead-ins as well as output line to receiver.
- Easy to install.
- Lead-ins attach to rear and are hidden from view



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sequential color disc exponent, indicated that he was delighted to hear of the tri-color tube, which he felt could be used in the CBS system at a much lower cost and without the addition of other receiver type tubes. He did not agree that the resolution problem was too acute. However, in an appearance before the Commission several weeks later, he described a modified system, using a sampling circuit based on the horizontal interlace principle, which provided a 525-line picture.

As soon as RCA had learned that CBS had outlined a method using sampling and horizontal interlace techniques, similar to what they had prescribed for their picture-dot interlace electronic system, headquarters issued a statement by RCA Lab Chief Dr. C. B. Jolliffe, which declared that: "At last CBS has admitted the weakness of its color television system. Because of lack of resolution in its system, CBS now states that it will now adopt the principles of sampling and horizontal interlace, developed by RCA.'

Shortly after the Jolliffe statement was released, CBS issued a biting re-Prepared by Adrian Murphy, ply.

NEW 500 KW. SUPER-POWER BEAM TRIODE

RADIO Corporation of America has recently announced the development of what is believed to be the world's most powerful electron tube, a high-vacuum "super-power beam triode" capable of 500 kw. continuous output and tested at one million watts İnput.

The new tube, RCA Type 5831, represents the first commercial attempt to build power output of this magnitude into a single tube. Despite its enormous power input capabilities, the new tube measures only 3834 inches in length and weighs only 135 pounds.

Applications of the new tube include high-power c.w. applications and international broadcast service. In unmodulated class C service, the 5831 has a maximum plate voltage rating of 16,000 volts, a maximum plate input of 650 kw. and a maximum plate dissipation of 150 kw. To provide 500 kw. out-put in unmodulated class C service, under typical operating conditions, the tube requires only about 900 watts grid drive. The tube can be operated with maximum rated plate voltage and plate input at frequencies throughout the standard broadcast band and much higher. The limitations for operation of the tube at high frequencies and at higher power have not been determined at present.

The outstanding structural feature of the tube centers around the "elec-tron heart," an array of 48 independent unit electron-optical systems arranged

RCA's new "Super-Power Beam Triode" which is capable of 500 kw. output and which has been tested at one million watts input, is shown with Dr. L. P. Garner, head of the company's Advance Development Lab at Lancaster, Pa., where the tube was developed.



cylindrically in the tube. Each of these systems consists of a filament in a slot in the beam-forming cylinder, grid rods, and the copper anode. Electrons leaving the emitting surface of the filament are beamed between a pair of grid rods to the anode by the focusing action of the beam-forming cylinder. Even though the grid may be positive, relatively few stray electrons strike it.

The individual filament and grid elcments of the unit triodes in the array are tungsten rods 8 inches long, sup-ported at both ends by means of knifeedge V-notch arrangements. Special feature of this design is the "panto-graphic mounting," consisting of a device with flexible spring-loaded fin-gers to which the filament rods are hooked. The pantographic device makes each filament strand and grid rod mechanically independent, and allows vertical movement without disturbing the precise alignments and spacings essential to effective electron optics. The filament strands are made of thoriated tungsten for economical opera-tion, as well as for high emission capabilities.

The internal elements of the tube, including the cylindrical beam-forming structure, anode, and filament, are concentrically arranged, adding to the high efficiency of the design, and permitting the use of a simplified internal water-cooling system for the beamforming cylinder and anode, with simple external hose connections. -30-

The "electron heart" of the new RCA Type 5831 tube consists of 48 independent unit electron-optical systems arranged cylindrically in the tube. Operator is shown mount-ing one of the 48 separate grid strands in place. The new tube is a water-cooled type.



RADIO & TELEVISION NEWS

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TYPE.	PRICE EA.	TYPE.	PRICE EA.	TYPE. PRI	ICE EA.	TIPE.	PRICE EA.	TYPE.	PRICE EA.	TYPE.	PRICE EA.
OA-1G	. 95	5.129	13.45	714	• 90	66B4	• 90	705A 706AY	1.55	955 956	• 55 • 55
01A FL=C1A	-45 3.95	5V-10 5W-1	1.07	9-3 10	•40 •55	7017	1.05	707A	14.00	957	• 45
143	. 60	523	- 80	10 ACORN	• 65	71A	. 75	707B	15.00	958 958A	+ 55
1A5GT C1B/3C31	+ 65 4. 85	523G	- 75	10 (VT-25A) 10E/146	+55 1.00	CRP72	- 95	708A 709A	3. 70 4. 75	959	- 55
1B4P	1.75	S-7	- 30 - 35	10T1	. 60	CYN72	1.75	710A	2.45	967/FG17	3.25
1821A/GL471A	2.55	EL-C6A	2.00	10Y (VT-25)	+45	RKR72 RKR73	+ 90 1, 25	713A 714AY	1.50	1005	- 24
1823	7.50	6A6	. 80	12A6GT	. 25	75	. 89	715B	9.75	1007	4.50
1B32/532A	1.85	6AB7	. 95	12A7 12AU7CT	+ 80	76 77	- 55	717A 7188V	- 85 15.00	1148	· 35
1842 1848	5.25 9.90	6AC7 6AF66	+ 90	12AU7	. 96	78	. 50	718EY	15.00	1203	- 55
EL IC	4.85	6AG5	1.20	12AX7	1.20	VR78	· 65	721A	3.75	1203A 1204	+ 65
105GT 106	. 65	6AH6	1.10	12BD6 12C8	1.20	80 FG-81-A	3,95	7224/2874	3,90	DG1295	9.95
1C 7G	- 85	6AK6	. 80	12F5GT	- 65	83V	. 90	723AB	14.95	1299/3D6 1613-SELECT	-45 6F6 -55
1D8GT 1E7GT	. 95 1. 95	6AL5	+ 95	12H6 12J56T	•40	89 89Y	. 40	724A 725A	4.25	1616	1.25
1154	+ 90	GAV6	+ 100	12J 7GT	.70	VR90	. 65	726A	12.50	1619	+ 35
166 1846	+ 65	6B4G	• 95	1268	• 65	VT90 (BRITIS) VR92	1) 2.00 .65	7268	13,50	1625	. 35
11(6G	. 80	687	• 75	12SA7	.73	FG95/DG1295	9.95	801	. 50	1626	. 35
11.4 11.06	• 50 • 75	6B8G	• 95	12807	. 75	VT98, REL5	14.95	801A 803	• 70 5. 25	1629	- 35 3.95
1LN5	• 65	6BA6 6BE6	+ 95	125F / 1256 7	• 00 • 65	100TH	11.50	804	8.95	1638	• 90
1N5GT 1P24	.75	6C4	.40	12SH7	•40	101/837	1.65	805 807	5,95	1641/RK60 1642	+ 65 + 50
105GT	.85	6C8G 6C21	1.05	12SJ7 12SK7	+ 73 + 60	102F FG105	3, 55	808	1.65	1852/6AC7	• 90
184 185	+ 55 - 70	606	19.20	12SK7GT	• 60	VR105	• 85	809	2.50	1853/6AB7 1960	+ 95
1T4	.75	6F5	• 85	12SL7GT 12SN7GT	+ 60	VU-111-8 114B	• 55 1, 20	813	2.90	1961/5324	1.85
2A3 2A7	1.05	6F6G	+ 60	12SQ7GT	• 60	1214	2.65	814	3.75	2050 2051	• 78
2AP1	4.75	6F8G	• 95	12SR7	+ 60	122A	2.65	815 826	2.85	UX6653	1.20
2B7	. 75	6060 6116	• 80 • 45	12X525-2AMP, TUN 12Z3	1,1,95 ,90	VT127A	2.95	8308	3.95	7193	• 35
2022/7192	3.25	6.16	- 90	13-1	. 35	VR150	50	832A	7.95	8011/VT90, B	A1113H2+55 3.25
2026	. 30	6.17GT 6.18G	• 70	14A7 14B6	+ 90 + 75	FG172	14.95	835/38111A	1.10	8013	1.25
2C26A 2C34	• 40	6K6GT	- 55	14F7	• 90	2058	1.45	836	1.35	8013A 8019	1.50
20.14	1.25	6K7	· 80	14117	+ 90	211 (VT-4-C) 2154 (VT5)	+ 60	838	3, 25	8020	3.25
2J21	10.45	61.60	1.35	14R7	• 90	CEP220	2.00	841	. 50	8025 9001	0.75
2.122	9,85	6L7	• B	15E	1.50	221A 227A	1.75	842 840	2.75	9002	• 45
2J26	8.45	6N7/GT	• 10	15R 16X879-2AMP. TUN	1,20	231D	1.20	851	39.00	9003	+ 40
2J31	9, 95	607	- 55	FG17/967	3.25	HX233A 250R	1,95	852 861	29,45	9006	-40
2132	12.85	6SA7	. 65	19 20-4 BALLAST	1.20	257A	3.00	861	• 45	38111A/ 83D	1.10
2,133	18,95	6SC7	. 75	21-2 BALLAST.	-45	268A 282B	2.95 4.25	865 866-10NTOR	2.55		
2J37	13.85	6SU76T 6SF7	• 70	REI 21	2.75	287A/722A	9.50	8664	1.30	YTAT. D	IODES
2J38 2J48	6+95 12-95	6SF5	• 65	RK24	1.75	304TL 304TU	1.75	869 8698	19.75	1891	p.e.
2.J61	24.50	6SG7 6SB7	- 65 - 40	24A	. 75	307A	4.25	872A	2.45	1N21A	• 00 • 95
2J62 2X2	14+95	6SH7GT	-40	VT-25-A/10	2.00 .55	316A	• 55	874 876	1,95	1N21B 1N22	1.20
2Y36	1.20	6SJ7 6SJ76T	+ 60 - 60	2525	- 73	3508	2.55	878	1.95	1N23	.80
3A4 3A4/4 3	+ 35	6SK7	+ 60	25260T 2526G	• 55 • 55	3540	14.95	879/2X2 902	• 55 3 50	IN23A 1897	· 85
3B7	-45	6SK7GT 6SL2CT	• 60	26	• 65	368AS/703A	4.95	923 (PHOTO)	1.35	1N29	• 80 • 85
3822	2.35	6SN7GT	- 85	27 2807	• 50 • 40	371A/VT62	. 95	930	1.00	1N51 (GE)	. 75
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3C31-C1B	4.85	6587 6U76	+ 80 + 85	GL34 BK34 /2034	1.50	417A	14.50				
3DP1	3.75	6V'6GT	.75	35/51	.60	434A 446A	3.40		U 3 I	UUI -	
3D6/1299	• 45	6W50 6X56T	· 80	351.6GT 25W4	.73	446B	1.55	_			
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3GP1	4.50	7-7-11 74.1/VYT	• 35	36	• 40	GL471A	2.55				
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5J23	4.76 14.25	717	• 90	65	1.25	703A/ 308AS 704A	3.95	L			



PARTS SHOW VISITORS: Be Sure to See Our Huge Display at Our LaSalle Street Show Rooms

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CBS vice-prexy, the text indicated that the RCA statement was a . . . "gross distortion of the facts."

To support its view that the pictures provided even by the original Goldmark approach were highly satisfactory, CBS offered the testimony of a parade of experts, one of whom was Dr. Jack W. Dunlap of a research organization in New York City. Reporting on a survey of doctors, who had viewed CBS color TV, he said that over 94 per-cent thought that the "over-all quality" was "excellent." Testimony offered by Oscar Katz, director of research for CBS, showed that only about 1.5 per-cent of close to 10,000 persons interviewed, found fault with the color wheel, a point which was disputed during a questioning period, some experts intimating that large-scale demonstrations do not provide a detailed evaluation of receiver operational features.

A BLANKET PREDICTION that the Commission would lift the freeze next fall was made by former FCC Commissioner E. K. Jett during a talk before a group of furniture dealers in Baltimore.

Jett, who also served as FCC's chief engineer for a long spell, declared that the ultra-high bands would be allocated, around September or October, for black and white service and that color may have to wait. None of the color systems demonstrated thus far, he added, showed promise of immediate adoption. It was his opinion that the Commission would not sanction any color system, unless it could be adapted to existing receivers at a very low cost.

RADIO AND TELEVISION will be used in quite a substantial manner by the Army and Air Force during their recruiting program, with about forty per-cent of the advertising budget set aside for the purpose. The news was revealed during testimony before the House Appropriations subcommittee. It is expected that over \$600,000 will be spent for broadcasts, in an attempt to procure over a quarter of a million volunteer enlistments.

The campaign is expected to begin during the summer.

THE LONG-AWAITED designation of a chief engineer for the FCC has been finally made, with Curtis Plummer chosen to fill the important post. Plummer, who has served as chief of the television broadcast division of the Commission, joined the FCC in 1940 as a radio inspector at Boston. Shortly after that appointment he became an assistant monitoring officer, and in 1941 came to Washington as an associate radio engineer. In '42 he became a radio engineer for the Commission, after which followed service in the nonstandard broadcast section of engineering and then the TV division.

Congratulations, old man. We know that you'll be eminently successful in this key assignment. . . . L. W.



PLAYBACK UNIT

Ampex Electric Corporation of San Carlos, California, has developed a new continuous playback machine utilizing twin-track magnetic tape. It has been designed especially for continuous high-fidelity music or voice reproductions for use in entertainment, educational work, and other services.

Known as the Model 450, the new unit utilizes the special *Ampex* magnetic playback head. The machine uses a synchronous tape drive mechanism in which no part revolves at excessive speed. There are no gears, belts, or rim drives, and all rotating parts run on permanently lubricated, sealed ball-bearings. Operation is entirely automatic and requires no attention other than turning on or off of electric power.

Two separate playback heads are used, one for each of the two recorded tracks on the tape. The machine automatically reverses by means of a special thyratron circuit after playing for approximately one hour and reverses again at the end of the second hour.

BLAST FILTER

Electro-Voice, Inc. of Buchanan, Michigan has developed and is manufacturing a blast filter unit which



stops wind and breath blasts and prevents pop in dynamic microphones.

The Model 335 blast filter is said to stop breath blasts even when the speaker works extremely close to the microphone. In outdoor pickup it completely eliminates any interference from wind rumble.

The unit is designed to be used with E-V Models 640, 635, and 605 moving coil dynamic microphones. The blast filter is small in size and can be easily installed on the microphone without adjustments.

RADIO CODE MACHINE

Ultradyne Electronics of Oswego, Oregon has announced production on a new radio code machine, the Model RCM-2.

Priced to fit the pocketbook of the individual user, this new machine

May, 1950 -

For additional information on any of the items described herein, readers are asked to write direct to the manufacturer. By mentioning RADIO & TELEVISION NEWS, the page, and the issue number, delay will be avoided.

teaches the code by photoelectrically keyed inked code practice tapes. The tape is pulled through the machine at



a speed which can be increased a little at a time until the desired code speed is attained.

Sending practice is facilitated by using one of the beginners' tapes that has extra spacing between characters. The operator can then imitate, as closely as possible, using a hand key, the correctly sent character sounded by the machine.

The self-contained unit comprises a tape-puller motor, take-up reel, rewind and phototube keying circuit and has three controls—keying, speed, and rewind. A complete selection of practice tapes is available for use with the machine.

TUBE TESTER

A mutual conductance tube tester which reads on a calibrated micromho scale as well as on a "Reject-Good" basis has been introduced by *Electronic Measurements Corp.* of 423 Broome Street, New York 13, New York.

The new tester, the Model 201, also checks tubes for gas content and checks all tubes from .75 volt to 117 filament volts as well as all loctal, octal, 9-prong miniature and subminia-



ture tubes. Testing can be done on cold cathode, magic eye, voltage regu-

lator tubes, and ballast resistors. Individual sockets for each type tube base eliminate possible errors.

The unit has a $4\frac{1}{2}$ inch meter and is available in either a sloping counter case style or in a portable model in an oak carrying case.

GOODELL AMPLIFIER

The Minnesota Eletronics Corporation of 97 East Fifth Street, St. Paul 1, Minnesota has introduced a new amplifier unit for home installation which will retail in the moderate price class.

Known as the *Goodell* "50," the new unit features continuously variable bass and treble control, 16 db. bass boost and bass attenuation at 40 c.p.s. and 16 db. treble boost and treble attenuation at 10,000 c.p.s. The amplifier has stabilized degenerative feedback, low internal generator impedance and the automatic switching of feedback resistors on each output tap.

A selector switch provides output impedances of 4, 8, 20, 250, and 500 ohms ganged to simultaneous switch-



ing of feedback resistors for optimum performance on each tap. A fourposition switch provides radio, 78 r.p.m. phono, LP phono, and television, with automatic equalization inserted for LP and 45 r.p.m. records. A socket is provided for a plug-in phono preamplifier.

Frequency response is said to be 20 to beyond 20,000 c.p.s. within 1.5 db. Power output is rated at 8 watts. A data sheet listing all of the features is available on the *Goodell* "50."

SOLDERING FLUXES

A line of fluxes, recently developed in England, is now being distributed in this country by *Newage International, Inc.* of 521 Fifth Avenue, New York 17, New York.

These T.S.F. fluxes are self-cleaning, remove oxides and film from ferrous and non-ferrous metals, and are impervious to grease, according to the



company. They do not corrode or rust and they do not leave corrosive residues.

The line includes various products for joining different types of metal. A consultant service on special metal joining problems is offered for those with difficult or unusual production requirements.

BOOM BRACKET KIT

Of interest to hams, recording enthusiasts, and commercial broadcasters is the new boom bracket kit recently announced by *Atlas Sound Corporation* of 1449 39th Street, Brooklyn 18, New York.

Designed as an all-purpose support for microphones, the "set screw" assembly makes it possible to cut down any tubular section so that the support bracket can be custom built to meet specific requirements. The microphone cable feeds through the entire support arm as well as the adjustable elbow mechanism. All parts of the bracket are finished in opalescent bronze enamel. The boom is 23" long, maximum.

SELENIUM RECTIFIERS

A new line of selenium rectifiers has been introduced recently by *Sarkes Tarzian Inc.* of Bloomington, Indiana. Designated the "Centre-Kooled" line,

these units have been designed for use in radio, television, or electronic equip-



ment. The center-cooling feature is provided by a special spacer between the cells to insure lower over-all operating temperatures by allowing air to reach the portions of the cells in which the current density is the greatest.

Other features include constant assembly pressure under all operating conditions, easily identified color coding, universal locating lug, and constantly high resistance path to ground under all atmospheric conditions.

Sixteen models are currently available and complete data and engineering information on these rectifiers is available from the company.

TUBE CHECKER

Two new tube testers, designed for testing of radio, television, mobile transmitting, and industrial electron tubes, have been introduced by *Syl*vania Electric Products Inc. of 500 Fifth Avenue, New York 18, New York. Available in either portable or bench models, the tester features an exclusive ohmmeter-type shorts and leakage test which indicates "Replace" or "Good" directly on the instrument's illuminated meter, direct meter indication for all other tests, an easy-to-



operate gas test, and a combined emission and transconductance test under dynamic operating conditions which takes relative tube life into account.

Twelve sockets provide for testing 4, 5, 6, 7, 8, and 9-pin tubes, octal and lock-in, miniatures, subminiature, acorn and hearing aid types, mobile and ruggedized tubes, and pilot lamps. Facilities for unannounced tube types are also included. Control settings are shown on a smooth, fast-running roller chart which is easily removable from the front panel for adding new tube settings as announced.

The portable unit has been designated the Type 220 while the counter model is Type 219. Both have identical electrical characteristics.

POWER SUPPLIES

A new series of electronically regulated and stabilized power supplies, utilizing a new type of direct-coupled amplifier to control a pair of thyratron rectifier tubes, is now in production at *Amplifier Corp. of America*, 398-2 Broadway, New York 13, N. Y.

Two separate series of 250 watt (output power) supplies are available. The standard series is stabilized against line changes of 90 to 130 volts within \pm .5% and load regulated within \pm .5% from no load to full load. The super series with a more sensitive



error control circuit is line stabilized and load regulated to well within \pm .1%.

Units are available in each series in

GREATEST TELEVISION VALUES RAD-EL-CO—OUTSTANDING

FAMOUS PHILMORE Prices Slashed on these fine TV





Finest 30-tube Television Receiver Similar to RCA Model 630TS with Automatic Gain Control, "Stand-ard" tuner and many other new features.

BUILD YOUR OWN

WIRED CHASSIS FOR CUSTOM INSTALLATIONS USE THE BEST Use Philmore Factory-wired chassis, ready to operate. MODEL CP-30D. 30-tube chassis for 16' CRT; this is the identical chassis used in the fine Commodore and Belvedere models below. Precision-built to laboratory standards. Design similar to RCA Model (30TS with latest improvements, including auto-matic gain control. Major components supplied by RCA. Voltage doubler provides 12,000 volts for clear, bright pictures on 16' CRT. Chassis completely wired and factory aligned. Includes 16' mounting bracket, wood block and ALL TUBES, except CRT. NET. S169.50 MODEL CP-30, 29-tube chassis for 10' or 12½' CRT. Similar to above but less voltage doubler. Includes mounting bracket for 10' or 12½' tube (please specify). Completely factory wired and aligned. With ALL TUBES, except picture tube NET. S164.50 FOR CUSTOM INSTALLATIONS USE THE BEST

BEAUTIFUL 16" CONSOLE \$27500

BLAUIIFUL 10⁻² CON More power, longer range, sharper pictures with this powerful 30-tube RCA type TV receiver. Excellent fringe reception, steady, clear pic-turcs. Picture tube is remov-able—no serews or bolts! These eabinet sets carry Philmore's famous CP-30D 30-tube chassis with AGC and new, improved built-in clea-tronic antenna that really works! All parts and subes carry standard 90-day war-wantee.

wantee. The COMMODORE Console

KOENIG "TELE-BEAMER" ROTATOR

ELECTRONIC SUPPLIES

VALUE IN TV ANTENNAS A B



E.-RT-52 LOW BAND

ME-48.—4-ft. steel	mast e	ktension.		
EACH	21.41	FIVE	AI	
The set of mon	2117	1211/12	4.77	C1 40
H BT-51 HIGH	BAND			
EACH	\$4.38	FIVE	AT	\$4.27
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F - RT-54 SIMPI	FY Y	TVPF		
EACH	\$3.50	FIVE	AT	

'UNIMAC'' CHIMNEY MOUNT THE REAL ÓÀ

Requires use of only one bolt on each of two units to lock-clamp the steel strapping and take up slack. Only one wrench required to assemble: takes mast 34' to 134' dia complete with 2 preassembled units, two 12-ft. bands and all hardware. Lots of 10......52.20 SET.

Lots of 10 JFD LIGHTNING ARRESTOR For 300-ohm line; protects TV set against lightning; requires no wire stripping; easy to mount. #AT-102 for regular twin lead, #AT-103 for heavy or tubular twin-lead. List \$2.25. EITHER TYPE. PER DOZ....\$14.70 EACH.......\$1.32

ALL-CHANNEL TV BOOSTERS MASCO MODEL MTB-13X Clearer, Sharper Reception . . . in weak signal areas. Extremely stable, with broad-band amplifier circuit boosting entire 6 mc band width of all TV channels. Broad and fine tuning. Shipping weight 8 Hos NET \$17.64 STANDARD CONTRACTOR OF THE OWNER OWNE TV BOOSTER **IV BOOSTER** The "Standard" pr e-a m p, made by nnfr. of the famous "Standard" tuner, boots sig-nal 10 times or more: high signal-to-noise ratio. Printed circuit assures trouble-free op-eration. Turning off booster automatically turns off TV set. Continuous tuning eliminates sitten from hi to locat SAVE \$\$ ON STANDARD REPLACEMENT PARTS

LOW COST GLARE FILTERS

TV TEST EQUIPMENT SAVE 50% WITH EICO **KITS AND INSTRUMENTS**

New push-pull wide range 5" oscilloscope. MODEL 425-Kit Complete nothing \$3995 else to buy....



SENSATIONAL SWEEP GENERATOR



500 KC to 228 MC: sweep width 0-30 MC, crystal

marker. EICO #360 Kit, less \$2995 crystal.....

MODEL 360 factory wired and aligned...... \$39.95

VACUUM TUBE



26 megohm input resistance AC-DC volts 0-5, 10, 100, 500, 1000; Resistance; 0-1000 ohms & 0-1, 10, 100, 1000 megohms. \$2395 EICO #221 Kit



Model 221 Factory wired & calibrated \$49.95 Hi-voltage probe #HVP-1 for #221..... 6.95

ALL EICO EQUIPMENT IN STOCK For complete description see EICO ad page 6 this issue

RADIO CITY PRODUCTS HIGH QUALITY-LOW PRICE

TV-90 'SCOPE WITH BUILT-IN SWEEP GENERATOR



SWEEP GENERATOR ONLY \$12245 Sevent the sevent strategy of the sevent sev



TV-50 MARKER ONLY \$4895

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Write for FREE catalog and flyers 219-R East 1st, TULSA 3, OKLA.



 LOW COSI GLARE HILLENS

 Relieves eve strain, sharpens, image, intensifies contrast, Optically correct. Available in either SMOKE or Birder (Dielase Spectrum)

 0. Birder (Dielase Spectrum)
 0.02

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TERMS: Money order or check with order; 25% de-posit for C.O.D., Postage extra.







. . \$2.40

WEETERS WIDEST SELECTION • BEST VALUE • HIGHEST QUALITY

SOLVE your hi-fidelity problems with

SINGLE UNIT TWEETERS

MODELS 4408, 4409-600 CYCLE TWEETERS: Recommended for highest quality reproduction systems requiring a law crossaver frequency. Cobra shaped horn results in perfect wide angle distribution. Frequency respanse 600 to 15,000 cycles. Model 4408 handles 6 watts and 4402 25 watts.

MODEL 4407 ADAPTER MOUNTS 4401 TWEFTER IN ANY 12" CONE UNIT: Converts any 12" cone speaker into a wide-ronge coaxial reproducer in a few minutes. Installation is extremely simple and results in a dual speaker occupying little more space than the original cone speaker. Complete with 4401 tweeter.

ER Ti ne ov xin in gte

MODEL 4401-2000 CYCLE TWEETER: An economical 6 watt unit for converting any good 10-15" cone speaker for extended response to 15,000 cycles. Wide Angle horn, compoct design and low price bring excellent high fidelity well within the populor price range.



DUAL TWEETERS

MODEL 4402, MODEL 4404: Madel 4402 reproduces to 15.000 cycles. Crossower at 2000 cps. Horizontal dispersion 100°, Vertical 50°. Handles 12 watts. Compact design mounts in any radio, phono, or speaker cabinet. Madel 4404 incorporates 4402 tweeter in handsome walnut cabinet complete with high-pass filter and high freauency volume control. Anyone can install.

CROSSOVER NETWORKS



MODEL 4405 HIGH PASS FILTER: An effective and economical unit for preventing lows reaching the tweeter unit. Cantains high frequency control to balance highs and lows. Cutoff frequency 2000 cycles.



MODEL 4410, 4420 LC CROSSOVER NETWORK: Genuine LC frequency dividers for segregoting highs and laws. Not to be confused with ordinary high-pass filters. Crossover frequencies: Model 4410 600 cycles, Model 4420 2000 cycles. Attenuator controls included and wired.

Write today for illustrated literature — address inquiries to Department A



the following voltages; 0 to 25 volts up to 10 amps; 0 to 50 volts up to 5 amps; 0 to 125 volts up to 2 amps; 0 to 250 volts at 1 amp; 0 to 500 volts at 500 ma. Each of the ten units has a continuously adjustable and smoothly controlled d.c. output to full voltage and current rating.

Full details are available from the company on request.

SUPERIOR'S TESTER

The Model TV-10 tube tester, which has been designed for checking AM, FM, and television tubes, is being marketed by *Superior Instruments Co.* of 227 Fulton Street, New York 7, New York.



The new unit uses self-cleaning lever action switches for individual element testing. Tubes having tapped filaments and filaments terminating in more than one pin can be tested because any of the pins may be placed in the neutral position when necessary.

Individual sockets are used for each type of tube, thus eliminating the customary practice of combination type sockets. The tester is housed in a hand-rubbed oak cabinet with a portable cover and measures $6'' \ge 11\frac{1}{2}'' \ge 12''$.

NEW V.T.V.M.

The Hickok Electrical Instrument Company of 10524 Dupont Avenue, Cleveland 8, Ohio has announced the availability of a new volt-ohm-capacity milliammeter, the Model 209-A, which has been designed especially to increase the speed and range in television servicing.

The new instrument measures any resistance, capacitance, voltage, or



current encountered in AM, FM, or TV receiver servicing. The unit features a 9" meter scale for easier readings of greater accuracy. The v.t.v.m. will measure resistance as low as 1/10 ohm and capacitance to $1\mu\mu$ fd. It permits peak-to-peak voltage measurements and contains a zero-center d.c. scale for increased accuracy in FM and TV servicing. The meter has a new a.c. range of 1200 volts and features flat frequency response to 300 mc.

Housed in a portable steel carrying case, the unit comes complete with low capacity, high frequency probes and all necessary test leads. $-\overline{30}-$

Within the Industry (Continued from page 32)

of RCA Victor . . . HARRY S. GOULD. formerly merchandising executive for The Magnavox Company, has joined John Meck Industries, Inc., as director of merchandising . . . ERNIE KOHLER has been added to the engineeringsales staff of the Oak Mfg. Co. of Chicago . . . GERALD C. SCHULTZ has been appointed chief electronics engineer for Gibbs Manufacturing & Research Corporation. He was formerly chief of the radar techniques unit at Air Materiel Command . . . The appointment of GEORGE P. LOHMAN as field sales representative of the RCA Victor Home Instrument Department in the northern part of the company's East Central Region has been recently announced . . . CURTIS R. HAMMOND is the new equipment sales manager of the Receiving Tube Division of Raytheon Manufacturing Company in Newton, Massachusetts . . . Freed Radio Corporation has named JOSHUA SIEGER to the post of vice-president in charge of engineering . . . JOHN F. HERBST has taken over the post of western regional manager for the receiver sales division of Allen B. Du Mont Laboratories, Inc. . . . DR. AN-TONIO R. RODRIGUEZ now heads the new ceramic laboratory for The Electrical Reactance Corporation . . . F. P. RICE has been named executive assistant to the manager of the tube division of Allen B. Du Mont Laboratories, Inc., while JAY L. FISHER has been named production control manager for the company's tube division . . Sentinel Radio Corp. has named DONALD G. HAINES to the post of chief electrical engineer . . . Hoffman Radio Corp. of Los Angeles has named two new vice-presidents, S. L. SPRAG-GINS to be in charge of plant operations, and JOHN A. RANKIN to serve as vice-president in charge of engineering . . . HARRY SEELEN is the new manager of the Lancaster Engineering Section of the RCA Tube Department. He succeeds Dr. Dayton Ulrey who retired recently.

* * *

MORHAN EXPORTING CORPORATION of 458 Broadway, New York, has been named export sales representative for *Pyramid Instrument Company*. The concern will handle all export work with the exception of Canada . . . W. B. PRAY has been appointed sales representative in the six New England states for *Operadio Manufacturing Company* of St. Charles, Ill. . . . PATTERSON & COMPANY of Dallas and MURPHY & COTA of Atlanta are han-





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RAKSCOPES ... D. C. AMPLIFIER

RAYONIC TUBES AND OTHER ELEC-

TRONIC TECHNICAL EQUIPMENT

dling the Alprodco, Inc., line of aluminum towers in their respective areas . . . The Circle "X" Antenna Corporation has named HARRISON-REYNOLDS **COMPANY** of Newton, Massachusetts, and **ROBERT MILSK** of Detroit to represent the company in those territories . . . BRUNO NEW YORK, INC., has been appointed distributor for Jerrold Electronics Corporation's line of multiple television antenna systems . . . The New York City, Long Island and Northern New Jersey territory will be covered by BEN JOSEPH on behalf of La Pointe-Plascomold Corporation's "Vee-D-X" antenna line . . . JACK C. GARDNER is the new district merchandiser for Bendix receivers in the Maryland and southern Delaware area ... JACK WEBER ASSOCIATES will cover the parts jobber accounts in New York State, New Jersey, Eastern Pennsylvania, Delaware, Maryland, and Washington, D. C. for Odegaard Manufacturing Company, producers of standoff insulators for TV installations . . . Starrett Television Corp. has appointed JOEL B. GOODMAN to act as its sales representative in Westchester and Putnam counties. New York . . . NICHOLAS J. LAUB is the new sales representative for Oxford Electric Corporation in the North and South Dakota and Minnesota territories . . . LEONARD D. ALLEN, INC., will handle the test equipment and transformer line of Freed Transformer Company, Inc., in the New York

ROGER BOWEN, formerly associated with the U.S. Signal Corps, has been



named to head the engineering department of Cannon Electric Development Company of Los Angeles.

During the war Mr. Bowen specialized in radar and general electronic

equipment, serving for awhile in England and later at Red Bank, New Jersey where he was in charge of standardization of all electro-mechanical devices. After the end of the war he joined the Signal Corps at Ft. Monmouth where he headed the electronic component research and development division.

D. Frank Jackson who has been acting chief engineer of the company will continue as chief assistant to Mr. Bowen.

> * *

FRED ELLINGER, who heads his own firm in the Chicago area, recently rounded out a quarter of a century as a sales representative, handling the lines of several well-known companies in the component field.

Mr. Ellinger entered the field in 1925 as an original partner in the R. F.Sparrow Company and later, in 1933, organized the Ellinger Sales Company. He has been active in "The Representatives" of Radio Parts Manufacturers, Inc., and has served both as president and chairman of the organization. He still serves on the board of governors of "The Reps." ÷ * *

"THE REPRESENTATIVES" of Radio Parts Manufacturers, Inc.'s New England Chapter recently elected officers for 1950.

Arthur E. Akeroyd was named president of the group with Walter T. Hannigan serving as vice-president. Robert A. Waters is the new secretary-treasurer.

In addition to the officers, chapter members named three delegates and alternates to attend the annual convention of the association to be held May 22nd in Chicago in conjunction with the Radio Parts Show. Mr. Ackeroyd, Mr. Hannigan and former chapter president Glenn M. Hathaway will represent the New England Chapter with Henry Lavin, W. B. Pray, and Henry P. Segal acting as alternates. -30-

George Sterling

(Continued from page 58)

cuits scheduled for inclusion. I wanted to be sure what I wrote was correct by personal check. My 'Radio Manual' had to be right!"

One of his first jobs at this time was to track down unlicensed radio transmitters used by gamblers to flash race results. Stations were concealed in buildings close to race tracks and the gamblers used high-powered glasses to observe race results which they reported by illegal radio. An automobile was fitted as a mobile detection and direction finding station. This and other measures soon made beating the races this way a hazardous gamble and the scientific pony players turned to personal sets that could be hidden on the body of a watcher at the track. Sterling promptly worked out his own "Dick Tracy" miniature detector which could be secreted in the hand. With this he could walk up to a perambulating radio bookie and give him the legal tap on the shoulder that said, 'Stop violating radio laws!'

In 1935 he was appointed Inspector in Charge of the 3rd Radio District. and shortly after was called to the main FCC office in Washington, D. C., where in June, 1937, he was appointed Assistant Chief of the Field Division, Engineering Department. With the need to observe clandestine radio operations becoming of the utmost importance, he was made Chief of the newly-created National Defense Operations Section, Field Division, on July 1, 1940, and given a sum of \$1,500,000 to organize and commence operations.

Immediately after Pearl Harbor, wartime protection of our country's interests started with the location and suppression of a trans-Atlantic transmitter which commenced operation in the German Embassy in Washington. This station was pin-pointed within

ORI)FB	TUB	ES D	rastically Reduced from 10 to 50% Nationally Advertised Brands
		Type Net Price 1A4P\$0.24 1A6	Type Net Price 5FP7 \$0.95 5J23 5.95	Type Net Price Type Net Price Type Net Price 7C4/1203A\$0.24 304TL\$1.29 837\$1.4 7E5/120139 32L7GT39 8412 00V/175A 32 24
		185/255	5W4	12A6
ARR	.UW!	1B32-532A 2.29 1C6	6AJ5	12A8GT
TEN TUBE SUPERHET RECEIV	ER	1D7G 19 1F4 24 1F5G 24	6F5GT	12J7GT
with crystal controlled local oscillator. Has provisi crystal channels between 108 to 112 MCPS con tubes and crystals but less dynamotor	ons for six plete with New \$7.95	1J6G	6K6G59 6L5G39 6L7G39	12SF5GT
Less Tubes and dynamotor but	New 3.95 New 3.95	1V	65F6GT	125 R7 GT . 29 700 A
RECEIVER—Easily Converted for Use in Cit Crystal Controlled Local Oscillator. Broad Band Pa E. (a. 1997) 26415 - 112587 - 2-12580	ixens Band ss20.7 MC 7 128D7	2V3G	6SG7	16 R .19 714 AY 5.95 Amperite 19
relays, crystals. Schematic furnished	YEW \$7.95	5BP4 2.95 5CP1 2.95 5D21 9.95	6U7G29 6Z7G39 6ZY5G29 WRTTE 1	30SPEC 829 6.95 VR150 (Vt67)
Westinghouse Auxiliary Relay Typ Style 423396 C-110 V 60 cycle 4 pole. Enclosed in glass case	New \$7.50		Miscella	neous SPECIALS
HERMETICALLY SEALED CHOP	CES	ID 6/APN 4 Scor R 7/APS 2 Recei	Used be, Excellent. \$29.50 iver-Indicator	New Used Ne
10 H. 100 M.A		R 78/APS-15 H cator. BC 1287 A Scope	eceiver-Indi- 	Manual for BC 312 & 342 J 1. Manual for SCR 269 G 2.5 FL 8 Filter
.51.51111 r	New \$0.49	SCR 522 Transc 150 MC	Scope 12.95 eiver 100 to 	Image: State
BC-605 InterphoneAmplifier		KC. MN 26 C or Y R. RA 10 DA Bacci		24.95 Compact aluminum case fully enclose
Easily converted to an ideal in- tercommunication set for office-		T26/APT2 Trans RT7/APN1 Trans APN 1 Complete	sceiver 5.95	9.95 24.50 24.50 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.55 3.553.55 3.55 3.55 3.55 3.553.55 3.55 3.55 3.55 3.553.55 3.55 3.55 3.55 3.55 3.553.55 3.55 3.553.55 3.55 3.55 3.55 3.55 3.55 3.553.55 3.55 3.553.55 3.55 3.553.55 3.55 3.553.55 3.55 3.553.55 3.55 3.553.55 3.55 3.553.55 3.553.55 3.553.55 3.553.553.55 3.55
Original—New		BD 71 6 Pos. Sw EE 8 Field Phone BC 347 Interphone	itchboard 9.95	12.95 and the second se
New		I-70 Tuning Mete AM 61 Indicator	Amplifier	89 phone amplifier. Complete with tube at - 9.50 shock mount, but less battery New \$3.
All necessary parts and instruc-		PE 237 Power St BC 461 Veeder F	loot Counter.	12.95 Motor—Universal Electric, 24 VDC, w also operate on 24 VAC Diameter 15% Length 2%, Shaft 14'x34', New S1.
tions to convert the above to AC operation with one remote sta-		BC 746—Bantam QST Jan. 1948	idenser 1.48 one watt foundation). Contains 2 crysta	BC 1291 —Control box contains motor rh ostat control rated 10 ohms at 3.88 amp Brand new with cord and plug-in ven
	20 340	tuning conden 11-12-13-76. F	sers, etc. Numbers lach Informatio	1-7-10- \$0.95 MC 385A—Headset AdapterNew \$0.4 n and Prices on Request
BC-604 Transmitter IVE 20 11 and 15 meters. Can be operated on 10 meters push button crystal. With all tubes and meter but less	-40 JVLU -10 channel 612 65	BC 639 Receiver RTA 1B Transce TA 2124 Transm	with RA 42 Rect	ifier SCR 584 Components R-132/TPS 10 Radar Receiver
dynamotor. Excellent Condition Crystals— Set of 80.	14.95	Pack SCR 269 Compare R 5/ARN 7 Com	ss Installation pass Installation	AN APRI Receiver and Tuning Units ASB7 Complete Radar Installation
BC-603—Companion receiver to above with tub dynamotor	es but less .Used \$17.50	MN 26 Compass I. L. S. Installat	Installation ion (R 89-BC733)	TS-251 Test Set BC 221 Freq; Meter
DYNAMOTORS DM-28—For BC-348 with Mount and	20 lbs: Ass't	radio parts: A \$25	CKAGE ^{.00} \$1.95	Used Net 512.95
Filter	Desetter T	BC 620	shample 20 to	BC-454
DM-36	27.8 MC Filament.	FM—13 tubes. M	letered, Plate and New \$14.95	BC-457
PE-206Used 2.75 PE-101New 2.75 PE-73	PE 97 Power type. Used—c	Supply for above	6-12 volt vibrator	BC-450-3 Receiver Remote Control I BC-442 Z.
DM-53	Used less FT 250 Moun	s tubes, vib. & cond t for both BC 620 ar	2.95 ad PE 97 New \$1.50	2 Transmitter Rack 1.50 Complete Command set as removed from aircraft 3 receivers—2 transmitters—Relay unit—contr
Used .95 (3 for \$2.00) BEAM INDICATORS	Brand new T two tuning	BC 223 ransmitter with all unit cases, spare t	three tuning units, ube carrying case.	boxes-mounting racks-plugs- modulator and dynamotors-crated. Set
182—5"	shock mou price of Set of 5 tul	nt and brace; but le	ss tubes at new low \$19.95 \$3.95	HS-23 Hi Imp Headsets
I 81-3"	Tuning units Cases at PE 125-12-	are available separ volt Vibrator Pack.	ately atEa. \$2.50 Ea	T-17D Carbon Mike
I 81Used 2.45	Spare paris 2 vibrators	kit for PE 125 c and 13 fuses in m	Used 8.95 ontaining 2 tubes; letal container with	T-30 Throat Mike
T-85/APT5 UHF TRANSMITTER operating over a frequency range of 300 to 1400 MCPC with a nominal output of from 10 to 30	Send for	r free 8-page	eillustrated	2 mfd. 4000 VDC. OIL FILLED
watts. Unit is equipped with 110 V 60 CPS fila- ment transformer; blower; lecher wire test fre- quency set, and 8 tubes-1-931A; 2-6AC7;	BU	JLLETIN N many excepti	o. 103 onal values	4 tor 10. 4 tor 10. 1 mfd. 6000 VDC. OIL FILLED 1. .25 mfd. 15000 VDC. OIL FILLED 4. .00025 mfd. 25000 VDC. OIL FILLED 4.
2-6AG7; 1-6L6G; 2-829B; 1-3C22 (GL522) (oscillator). New in original box with Operating CRO EA	Comments of	1000		4 mid, 1500 VDC, OIL FILLED 2 2 mid 600 VDC, OIL FILLED
All shipments FOR Chicago 2007 Denosit apprint	AKK	UW SAL	ES, Inc.	1 mfd, 600 VDC. OIL FILLED
on all orders. Minimum order accepted—\$5.00. Illinois residents, please add regular sales tax to your remittance.	1712-14 5.	Michigan Ave., PHONE: HArrison	Chicago 16, 11 7-9374	.1x.1x.1-1200 VDC. OIL FILLED 2 for 1. 50 mmfd-5KV-5 Amp. Vacuum Cond 1.
May, 1950				****

IN DETROIT IT'S AAR	ON
BC-348 RECEIVER—used but in good condition —less tubes	\$49.95
(Will supply tubes at cost) SCR-522 TRANS/CEIVER-used-less tubes (Will supply tubes at cost) T-26/APT-2 RADAR SET_80/115V 1 phase- 400 2600 gmb 2 35 VDC 660.day.44 gen	29.95
tains following tubes:(2) 5R4 (2) 6AC (1) 2X2 (1) 807 (1) 6AG7. New with manual I-198-A SIGNAL GENERATOR-6-15 MC-New. ABK OR BC-966-A IFF RCVR. & XMTRwith	14.95 19.95
14 tubes & dynamotor—practically new—nevel been used (DYNAMOTOR same as 4 speed being adver- tised for \$5.95—we give you 14 tubes & other useful items for the same price)	5.95
R1/ARR-1 HOMING RCVRJan. 48 RADI(NEWS tells how to convert this rcvr. to Hi Freq. for use with your present rcvrnevel been used-has 4 acorn tubes-12 or 24	
VDC operation MINEAPOLIS-HONEYWELL AMPLIFIER—con tains following tubes:—(2)7C5 (1)7F7 (1)7Y4 —,1—,1 MFD 1000 VDC cond.—7 resistors— 2 fuses & fuse holders—400 cv. xfmr.	. 5.95 . 1.95
BC-1141 AMPLIFIER—used with BC-625 Mine Detector new with following tubes:(2)1Nt (1)105 SERVO AMPLIFIER-new-less tubes-contain	5,95
ets—8 relays—hundreds of other parts REMOTE INDICATOR BC-1153-A—with 5FP7 CI tube BC-741-A Indicator—Contains (12)6AC7 (1)6SN3	3,95 5,95
(1)6L6 (1)2X2 (1)CR1812 P7-5 neon lights- numerous amounts of other misc, parts-have been exposed to weather but can be cleaned up	19,95
SCR-521 (ASB) with 7 tubes less 5" CR tube & shield-makes excellent scope-brand new in a beautiful 30" x 11" x 10" wood chest R-29/AP5-2C-37 tubes-including (1) 5FP7 (1)	9.95
2AP1 CR tube-with blower motor etcNew in crates RADIO MODULATOR BC-188-A-contains Weston #506 200 MA DC meter-used with BC-187	39.95
ATB RADIO XMTR. CRV-52233_new in cartom -with dual plug in tuning units for 3.0-9.05 MC-contains following tubes-(1)1625 (1 VR-150 (2)6N7 (2)815) 29.95
ALTIMETER RT/APN-1-430 MC Revrcontain 14 tubes-3 1 meg. precision resistor-dyna motor numerous other valuable parts-new. BC-645 Trans/ceiver-430-500 MC-can be con verted to Citizens & Ham bands-new fit	5 7.95
cartons BC-646 CONTROL BOX for BC-645. New BC-733 LOCALIZER RCVR108-110 MC tubes10 crystals. New in cartons	12.95 1.00 6.95
BM-53-A DYNAMOTOR-used with Bc-103 AN/CRW-2 VHF RCVRcontains the followinf- (1)6SN7 (1)655 (1)6SC7 (3)6SL7-dynamoto -numerous amount of resistors & cond BC-229 RECEIVER-used with tubes.	5.95
BC-430 XMTR.—used with tubes both for C-439 COIL UNIT for BC-229—(201-398 KC & 2500-4700 KC). New WERE SUBLY BE 157-R. Contribution	7.95 .69
(1) 2 V vibrator & (1) 6 V vibrator (2) rectifiers (1) 4" spkr.—all for	7.95 1.00
RESISTOR & CONDENSER KITS—contains 44 resistors—various ohmage & wattage values— 20 conds., mica, ceramic and oil filled. New. VOLUME CONTROL KITS—contains 10 controls— VOLUME CONTROL KITS—contains 10 controls—	2.95
From 2010 to 250,000 of mNS=2 w. New, COLL SET & CRYSTALS for RX-TX BC-721-B- on freq. of 5500 KC (5500 & 5955 xtals. ant. & tank colls)	.79
350-25 VDC. New COIL ASS'Y BOX-contains variable & fixed condchoke-resistors-coils-out of BC-31 Revr. New PUTTERLY COND. With silver plated loop of	59 1 2 . 1.75
tached_approx. 1" x 1" x 1". New 98-156 M STANDARD BROADCAST TUNING CONDENSE	38 .38 .49
	69 . 1.95 . 1.00
CERAMIC COLL FORMS-2" dia71/4" L. 41/2 L or 21/2" L 30 MC SLUG TUNED IF COLL New	n .50 .25 .97
274-N COMMAND EQUIPMENT	
0-1 ma Thermocouple R.F. meter\$ BC-453 Receiver—used but in good cond	2.25 12.95
BC-454 Receiver-used but in good cond. BC-456 Modulator. Used \$1.95,New BC-457 Transmitter Used	2.50
BC-458 Transmitter. Used BC-459 Transmitter. New in cartons	6.95 17.95
BC-696 Transmitter. New in cartons MC-215 Control Cable. 77" long DM-33-A Dynamotor for BC-456	24.00 1.25 1.00
BC-450-A Control Box for receivers	1.50
 & PL-26 plugs. 8 ft. VT-127 Ceramic Tube sockets J-38 BREAK-IN-TELEPHONE KEY—on Bakelit base—444" x 3"—has binding post—circuit closing switch=1/16" silver kening context 	.75 .3/1.00 t
etc. CX-496-TP5-2 CORD—with AN-3052-10 & AN 3108-20-4P amphenol plugs & connectors New	
 HONE & WHETSTONE_for shops_fishing_hund ing(handle is cork burnisher) 4" x 1/2": 1/2" 3 for 25c. 4 VDC BELLS_New_use for alarm systems- wire two in series to work on 110VAC 	> x
3 SECTION CERAMIC WAFER SWITCH. New- multi-contact—6 position STATIC PRESURE SELECTOR VALVE SWITCH- Aircraft type. Part #AN5831-1. New- NAMULO WIFE SHEADEETS NEW 10.10	69
REVERSIBLE MOTOR—Bendix M0-5-B-28 VAC DC—7000 RFM—approx. 1" dia. x 21/2" long ELECTRIC MOTOR CORP.—Type 1-6 VDC with worm gear	z 1.95
 24 VAC-DC REVENSIBLE MOTOR-40 lbs torque 1.4 amps-3 RPM PERMANENT MAGNETIC MOTOR-±506857- 27,5 VDC 10.000 RPM-made by Delco 24 VDC SWITCHING MOTOR-simple conver 	6.95 3.75
sion to 115 VAC-DC-CONVERSION DATA INCLUDED-New DUNMORE SERIES WOUND MOTOR-Interval ometer Type B-4-Gear Ratio 40.1-New.	2.95
electronic equipment. All items USED, unless otherwise stated.	or used lichigan
including postage C.O.D.	Jaiance

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hours after its first attempts to contact Germany. It was actually closed down before it had succeeded in establishing the desired contact!

The increasing size and importance of this work resulted in his promotion to Assistant Chief Engineer and Chief, Radio Intelligence Division, on June 1, 1942. On December 19, 1945, he was made Assistant Chief Engineer in Charge of Field and Research Branch; and on May 1, 1947, he was promoted to Chief Engineer of the Commission.

He is probably proudest of his work as organizer and wartime head of RID. Testifying as head of RID before a Congressional Committee in 1944, he said, "I readily admit that this country has been practically free of espionage radio activity and I am extremely proud of that fact. . . . The success of RID, like the success of any police organization, is properly measured, not by the amount of crime which is detected, but rather by the organization's success in preventing illegal conduct and maintaining order." The final score: no enemy spy ever successfully established clandestine radio operations from the U.S.!

George Sterling today is still an active ham operating as W3DF. He has two transmitters, a 600 watt phone rig for 75, 20, and 10 meters and a 400 watt c.w. rig, mostly on 40. Receivers are an SX-42 and an SX-71. Separate dipoles are used on each band, with a three element rotary beam for 10.

The family car includes a ten meter ham rig; it also carries a v.h.f. job on 152.03 mc. used under the "Common Carrier Service."

He also has such things as a 16 inch TV set for family use (FCC Commissioners are not permitted to touch its insides); a 10 inch TV set with a juryrigged color adaptor; a commercial color receiver on temporary loan for test by the Commissioner and his friends, non-technical as well as technical; a basement workshop which he uses as much as a hurried, crowded official life will permit; and the walls of his ham-shack den are covered with photographs, diplomas, QSL cards and other mementos of a long association with the radio industry.

Commissioner Sterling is the author of the well-known and authoritative "Radio Manual," a text and reference work for radio operators, engineers, and amateurs. First published in 1928, it has been kept up-to-date and the postwar edition has just become available.

Sterling is also a Fellow of the Institute of Radio Engineers and on the Board of Directors of the Veteran Wireless Operator's Association. The Marconi Medal for Public Service was recently awarded by the VWOA. He enjoys attending on-the-air meetings of the Old Old Timer's Club, an amateur organization whose members must have been hams at least forty years ago and must refer to "wireless" instead of "radio" or "electronics" in their sessions. He is an honorary member of the Baltimore Amateur Communications Society; and the Quarter



Century Wireless Association of New York: the Chicagoland Mobile Club; and the Puerto Rico Amateur Radio Club.

As Chief Engineer of FCC, he participated in preparatory work and assisted at the Atlantic City Telecommunications Conference of 1947. In late 1947, he was Chairman of the Broadcast Engineering Conference under the North American Regional Broadcasting Agreement (NARBA), a most vital safeguard to American broadcasters. Commissioner Sterling was co-chairman of the U.S. delegation at the 1948-49 Mexico City International Short Wave Broadcasting Conference, which worked out the Mexican Plan for high-frequency broadcasting.

Queried as to the future prospects of Amateur Radio in the face of tremendous competition for all available frequencies, he made this statement for RADIO & TELEVISION NEWS' amateur readers:

"There are not enough channels altogether to satisfy more than a small part of the demand. This is particularly true in the high frequency region, which includes the 3.5, 7, 14, and the proposed 21 mc. ham bands. Those who are assigned frequency space must make the fullest possible use of it; and in the public interest. Each of the various services, of which the Amateur Radio Service is one, must justify the precedence and scope of its claim with relation to all others. Naturally, this must be in the terms of the present and the future worth of the service concerned.

"We know that the frequency allocation picture is unsettled and that there are powerful groups who would divert amateur frequencies to other uses. The amateurs should consider how to better justify and support their service—both within this country and in the international field. Under the conditions of today, there is indeed cause of concern. Amateurs must gird their loins for the battle of 1952 scheduled to take place in Buenos Aires, Argentina!"

-30-

PRSMA CHANGES NAME

THE country's oldest service technicians' association, the Philadelphia Radio Service Men's Association, has recently announced that the organization will henceforth be known as the Electronic Technicians' Guild.

The new name was adopted to reflect the new and wider scope of the organization than was implied by the former designation. The Guild membership will now include radio and television dealers, distributors, distributor sales personnel, and independent radio and television service shop owners and their employees.

In addition to renaming the organization the Guild Committee Chairman Karl Vogelsang presented a new plan of operation for the approval of the membership which will increase membership benefits and services. -50-





Model TVG-1 Net Price \$220.00

The ONE generator for accurate complete TV servicing. Look at these features:

Built-In Marker Generator. On fundamentals from 4 to 54 Mc. Harmonics readily usable to 216 Mc. Three ranges, accurately calibrated dial.

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Thanks to an amazing JACKSON discovery, this fine laborotory-type instrument gives you either wide band or high sensitivity. Check these features against any other oscilloscope.

Wide Band Amplifier. Vertical amplifier flat within plus or minus 1.5 db from 20 cycles to 4½ Mc. Readily usable to full 5 Mcl Such band width is essential for accurate picture of TV sync pulses. High Sensitivity. Vertical sensitivity .018 RMS volts-per-inch. Ideal for proper analysis of low voltage signals. Band width on High Sensitivity ranges is 20 cycles to 100 Kc.

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International Short-Wave (Continued from page 54)

Listeners are reminded that Brazilian stations likely will change schedules from around May 1 when that country returns to Standard Time from Summer Time and they should adjust the listed programs to conform to the new times which are, or will be in effect very shortly.

Our best wishes go to all radio outlets in our neighboring country to the South—Brazil!

New Allocations

Regarding the possibility of new high frequency allocations, I have received this information, on request, from T. G. Slewie, Secretary of the Federal Communications Commission:

The Mexico City Conference, which adjourned in April 1949, formulated one phase of a plan for high frequency broadcasting for the June season, sunspot number 70. This phase has not been accepted by the United States. It will be subject to modification, along with addition of phases formulated by the Technical Plan Committee at Paris in the fall of 1949, when the work of the Mexico City Conference is resumed in Florence, Italy, this spring. (The Florence meeting was to have begun April 1. -KRB) Those phases of a plan approved at Florence then will become a part of the draft International Frequency List covering all servicesexcept amateur-which have been allocated frequencies below 27,500 kc. in the Atlantic City Radio Regulations. The whole draft List will be considered by the Extraordinary Administrative Radio Conference, scheduled to convene September 1 at Geneva, Switzerland.

The date and the method for implementing any List approved by the Extraordinary Conference have not been determined as yet-but these matters will be considered at that conference.

"It appears at this time that no one service, such as high frequency broadcasting, will be able to implement its portion of the List in advance of the other services because of the interrelated problems of all of the services," Mr. Slewie points out. "It appears doubtful, also, that the Extraordinary Conference can complete its work in time for implementation of the new List to begin during this calendar year."

Portuguese List

Revised call letters for Portuguese stations, from ITU frequency List Supplements, are—Emissora Nacio-nal, Lisbon, 10 kw., CSA38, 6.360; CSA21, 6.374; CSA22, 7.260; CSA23, 9.635; CSA24, 9.670; CSA25, 9.680; CSA26, 9.727; CSA27, 9.740; CSA28, 9.940; CSA29, 11.027; CSA30, 11.040; CSA31, 11.840; CSA32, 11.995; CSA33,

12.749; CSA39, 15.100; CSA35, 15.110; CSA36, 15.120; CSA34, 15.160; CSA37, 15.320. Private Stations-CSB53, 3.960, 10 kw., Lisbon; CSB56, 6.003, 200 w., Lisbon; CSB52, 6.154, 2 kw., Lisbon; CSB54, 6.185, 350 w., Lisbon; CSB55, 9.650, 2 kw., Lisbon; CSB57, 12.130, 2 kw., Lisbon; CSB51, 12.865, 500 w., Parede. Ponta Delgada, Azores, 1 kw., CSA93, 4.845; CSA91, 7.018; CSA92, 11.090. (Legge, N. Y.)

World Listening Guide O. Lund-Johansen, Copenhagen, Denmark, who compiles World Radio Handbook, informs me he is preparing a World Listening Guide. He says:

'In order to promote interest in listening to broadcasts from foreign countries-particularly on short-wave -and to give short-wave listeners all over the world a chance to share each other's experiences, World Radio Handbook has planned the production of a world listening guide. In this booklet, short-wave listeners the world over will be writing about their experiences, practical handling of the receiver, giving reception reports, technical tips about aerials, and many other matters."

I hope soon to have further information on this project.

Mr. Johansen's QRA is 1, Lindorffsalle, Hellerup, Copenhagen, Denmark.

* *

Club Notes England—Frank Baldwin is now General Manager for the International Short Wave League, and Mrs. Baldwin operates the club's QSL Bureau.

This Month's Schedules

(NOTE: At this season many shortwave stations will be going on Summer Time; in such cases, schedules may be changed to one hour earlier than listed herein.—KRB)

Albania-ZAA, 7.852, Tirana, is now scheduled daily 1230-1600. (Radio Sweden)

Anglo-Egyptian Sudan-Verification and schedules received from Omdurman Broadcasting Station, Khartoum, listed frequencies of 5.000 (temporary), 9.770, and 572.5 kc. Station

Leader, composer, and arranger Radames Gnatalli conducts Radio Nacional's orchestra.



RADIO & TELEVISION NEWS





has English on Fridays 1230-1300.

Angola—CR6RL, Radio Clube An-gola, Luanda, is testing on a *new* channel of 5.096 at 1320-1600, according to Stephan, South Africa; is heard better than dual CR6RN on 9.469, Stephan reports. A station heard on 7.590 in Portuguese to closing around 1510 has been identified as Silva Porto, Radio Clube DB, CR6RO; however, according to its QSL card, CR6RO is on 7.582. (Radio Sweden) Reports on the new Nova Lisboa outlet on 11.925 should be sent to P. O. Box 125, Nova Lisboa, Angola, according to Radio Australia. This one has been heard by Stark, Texas, to 1700 when closed with "A Portuguesa." Radio Sweden reports the outlet heard in Portuguese around 1300-1640, identifying as Radio Clube de Huambo, Nova Lisboa.

Argentina-LRM, measured 6.1806. Mendoza, noted 2320 with music. (Treibel, Washington)

Australia—Current Radio Australia transmissions to North America are 2330-0045 to West Coast, VLC9, 17.84, VLA6, 15.20, VLG6, 15.32; 0700-0900 to East Coast, VLC5, 9.54; 0900-1000 to Central and Mountain States, VLC5, 9.54; 1000-1115 to West Coast, VLC5, 9.54, VLA8, 11.81; 1643-1815 to East Coast, VLA6, 15.22.

Austria-Blue Danube Network, Salzburg, logged 0900 on approximately 9.530, apparently is still moving around the 31-m. band. (Pearce, England)

Belgian Congo-OQ2AB, Radio Elisabeth, Elizabethville, operates on 7.150 and 11.900 on Sunday 0800-1000; OQ2AC, Radio College, Elizabethville, operates on 3.390, 4.980, 7.200 daily 1130-1215, Sunday 0200-0300. (Nordh, Sweden, via Radio Sweden)

Brazil-PRL5, 11.95, Rio de Janeiro. good level in Newfoundland from 1630 to after 1730. (Peddle)

British Guiana-Grove, Ill., says ZFY, 5.985, Georgetown, has announced expansion of its service to 0545-1145, 1445-2045.

Cameroons-FIA6, 9.145, is very poor at 1430-1500. (Peddle, Newfoundland)

Canada—Anyone interested in the attractive monthly program booklet issued by CBC's International Service may get on the mailing list by sending his request to CBC International Service, P. O. Box 7000, Montreal, Province of Quebec, Canada. For each of the main areas-Europe and Latin America (including the Caribbean)-CBC publishes a separate booklet each month. Listeners should specify which booklet they want—European or Latin American. Latest schedules are -to Europe, CKNC, 17.82, 0915-1615 (Sunday from 1035); CKCX, 15.19, 0915-1128 (Saturday from 1020); CKCS, 15.32, 1130-1855; CHOL, 11.72, 1630-1855. To Caribbean and Latin America—CKRX, 11.76, CKCX, 15.19, 1910-2200; CKCS, 15.32, 2200-2230. To Australasia (United Nations Service) --Mon.-Fri., 2320-2400, CKLX, 15.09,

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Square 4" PM Alnico V 1 oz.

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20	14154**	rnı.	113 v. sec. 000 voi 0.5 v 5 Amp. 5 v. 2.5 Amp	pr.50 caci
80	MA.	PRI.	115V. sec. 700 VCT 6.3V 3 Amp. 5V. 2.5 Amp	2.35 each
100	MA.	PRI.	115V. sec. 600 VCT 6.3V 3 Amp. 5V. 2.5 Amp.	2.95 each
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50	MA.	PRI.	110-120V. sec. 600 VCT 6.3V. 3 Amp	2.25 each
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CHOL, 11.72. To Australasia (Sunday only)-0340-0530, CHOL, 11.72, CKLO, 9.63.

Cape Verde Island-CR4AA, 5.92, heard in Pennsylvania 1650 with music, heavy CWQRM. (Slutter) *

Ceylon-Radio Colombo, 4.900, noted 1115 following BBC news to give local time as 9:45 p.m.; had music 1130-1200 closedown. (Pearce, England)

Cyprus-ZJM7, 11.72, Limassol, fair 1200-1230 in Newfoundland. (Peddle) ZJM3, 4.06, heard in Stockholm around 1200. (Radio Sweden)

Czechoslovakia-Balbi, Calif., reports OLR3A, 9.55, Prague, from 0000; is QRM'd at 0030 by Paris opening on this channel; also heard in Texas.

OLR4B, 11.76, Prague, noted 0120 with Spanish news at dictation speed. (Treibel, Washington)

Improved signals are widely reported in the 1900-2000 daily beam for North America over OLR4A, 11.84; news at start.

Denmark-OZU, 7.26, noted with relay of Home Service to closing 1730; news in Danish 1300, 1600; QSL card received. (Pearce, England) Copenhagen lists current s.w. schedules as 2100-2230, OZF, 9.52, to Greenland and North America (may be off Sunday); 0500-0600 on Tuesday, Thursday, Saturday, OZH, 15.165, to the East; 19002000 on Monday, Wednesday, Friday on OZH, 15.165, to South America; daily 0040-1730 on OZU, 7.26, with Home Service for Europe; transmitter is listed as located at Herstedvester, with 50 kw. (Klein, N. Y.)

Ecuador-HC2CA, measured 6.8852, Salinas, noted 2130 with Latin American music (Treibel, Washington)

Ethiopia-A letter from ETAA, Addis Ababa, verified reception of the 19-m. outlet; gave English program schedule as daily 1020-1100, Sunday 0945-1115; no frequency was listed but said was operating in the 19-, 31-, and 49-m. bands. The 19-m. outlet appears to be on approximately 15.032, although most overseas sources list it around 15.050 to 15.075. Radio Sweden says ETA, 9.62, is audible in Sweden around 0600.

Finland-Verie from Helsinki listed OIX1, 6.120; OIX2, 9.555; OIX4, 15.190; OIX5, 17.800. Stated daily program times for 15.190 as 2200-0000, 0700-0800, 1145-1245, 1600-1700; card had photo of the 100 kw. transmitters. (Hankins, Pa.) OIX5, 17.80, noted 0715-0730 with news for North America. (Peddle, Newfoundland)

Formosa-See Taiwan.

France-Paris noted signing off on 6.145 recently 2445. (Stark, Texas)

(Continued on page 159)

PRECISION SPONSORS SERVICE MEETINGS

DURING 1950 television technicians throughout the country will have the opportunity of participating in a series of lecture-discussions being sponsored by Precision Apparatus Co., Inc., of Elmhurst, New York.

Designed to assist the television technician solve his servicing problems, this series of practical lecture-discussions has been tailor-made and is being presented under the general title of "Tele-vision Circuitry, Video I.F. Alignment Practice, and Troubleshooting." The sessions are being conducted by R. G. Middleton, Senior Engineer of the Precision Design and Application Engineering Section.

Each lecture-discussion is broken

down into two periods. The first half of the session is devoted to a circuit analysis of the television high voltage power supply and how it works, followed by an open floor question and answer period. After a short intermission the second half of the program covers video i.f. alignment procedures including troubleshooting, short cuts, and efficient test equipment applica-tions. An actual TV alignment demonstration follows the lecture with the lastest test equipment being featured.

The company's plans now call for lecture-discussions in every primary TV area in the United States during 1950. -30-

R. G. Middleton addresses his audience of over 300 television technicians who gathered recently at Queens Village, Long Island, for the first of a series of nationwide lecture-discussions sponsored by Precision Apparatus Co. and its distributors.



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6 tubes: 2—6SJ7, 2— 6L6, 1—6SC7, 1—5Y3.

2 mike, 1 phono input.

Separate bass and

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Attractive heavy steel chassis and

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for hum level.

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Fan Type Antenna (Continued from page 67)

and therefore, all radiation is cancelled. However, when the conductors are bent out to form a dipole, the cancellation can no longer take place and the dipole antenna will either radiate or receive.

This dipole has an impedance of approximately 73 ohms. Now if the diameter of the dipole is increased after a right angle bend so that it constantly increases as we get farther from the point of the bend, a three dimensional cone is formed. This cone will still radiate or receive energy. However, its impedance characteristics are different. In this case, as the transmission line has been fanned out, the diameter of the conductors has been increased. The result is that the impedance remains constant over a wide band of frequencies. In other words, by the use of conical elements, a broadband antenna has been secured. In summarizing the previous two paragraphs, it can be seen that by veeing an antenna forward very narrow horizontal lobes and patterns can be obtained over a relatively wide band of frequencies.

In the case of the conical antenna, by the use of this uniform impedance conical element, it will work over a wide band of frequencies with a low standing wave ratio.

It must be borne in mind that the above description pertains only to driven or receiving elements and does not pertain to reflectors. The requirements of a reflector are such that it must re-radiate energy to the driven element in such a relationship that reinforcement or addition occurs. To fulfill these requirements, a straight reflector is as effective as a conical reflector or a fan shaped reflector provided that it is cut and spaced properly. In the laboratories, the above considerations were given careful thought. It was obviously not feasible to make a three dimensional sheet metal cone or even a wire cage type cone element. Previous conical antennas had simulated this cone by running two dipoles from a common apex producing a cone in two dimensions. It was further observed that in the case of a "V" antenna as the diameter of the legs of the "V" was increased, the antenna would become increasingly uni-directional in the direction of the enclosed part of the "V." It was then reasoned that if the conical element which operated effectively over the entire low band were bent forward so that it formed a "V" on the high band, it would provide an efficient all channel antenna. However, two shortcomings were observed. Veeing forward the antenna reduced the amount of energy intercepted on the low band because the distance across the front was decreased and secondly the two element conical antenna did not have an effective diameter large enough to pro-





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Crystal marker oscillator with variable amplitude. Covers all TV and FM alignment frequencies be-tween 500 kc. and 228 mc. Sweepwidth variable from 0-30 mc., with mechanical inductive sweep. Extremely wide sweepwidth allows gain compari-son of adjacent RF TV Channels. Provides for injection of external signal generator marker. Phasing control included. Vernier dial calibrated in frequencies. Complete with tubes (6X56T, 12AUT, 2-6C4). In sturdy steel case. Less Crys-tal. Size: $10 \times 8 \times 634^{\circ}$. 110-125 v., 60 cycle AC. Shog, wt., 12 lbs. IN KIT FORM. No. 32P24370: Concord's Price \$29.95

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May, 1950



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EDLIE ELECTRONICS, INC. 154 GREENWICH ST. NEW YORK 6, N. Y. duce a workable front-to-back ratio on the high band. A third element was added. This element overcame these two shortcomings. On the low band, the third element meant that more energy was being intercepted and therefore the narrowing of the space aperture was overcome. On the high band, the effective diameter of the legs of the "V" was increased so that a high frontto-back ratio was realized. As a matter of fact the same front-to-back ratio was realized on the high band after the inclusion of the third element, with or without reflectors. In addition, to these two improvements, it was found that the third rod made the elements more closely resemble a three dimensional cone and a more uniform impedance was realized. Therefore, a better match was made to the transmission line over the entire range of frequencies. The impedance was found to lie in the neighborhood of 150 ohms so that good matches were available at 72 ohms and at 300 ohms. The finished antenna is shown in Fig. 1. Horizontal polar diagrams at Channels 2 and 7 are shown on Fig. 5. Gain curves for the single, double, and four-bay arrays are shown graphically and tabularly in Fig. 6.

-30-

AN ELECTRICAL LOCK By WILLIAM K. COOPER

THERE is an easy way to prevent meter movements from being damaged by jars and rough handling when making outside scrvice calls. The tester is often bounced around in the car and accidentally bumped into things when carried on such trips. This can damage the bearings of the movement due to the sudden jars and the needle hitting the "stop" at the low end of the scale. Meters that are lightly damped need the most precautions in handling since they are most susceptible to damage caused by jars.

After having a meter damaged in this way, I found a simple means of preventing this trouble by simply using an "on-off" toggle switch. Drill a hole for it on the panel as near as practicable to the meter. If there is no room on the panel, mount it on the meter case. Before drilling a hole in the panel, the meter must be disconnected and taken out of the tester to prevent damage from vibration. Care should be taken that no circuit components are damaged by the drill. When the switch is mounted, the meter is replaced in the panel.

The switch is wired directly across the terminals of the meter itself and the meter reconnected to its circuit as before. With this switch "on" the moving coil is shorted. Then any motion of this coil as the tester is carried generates a current of such direction that it opposes the coil motion. This is in accordance with Lenz's law, remember? Thus there is dynamic braking of the meter movement. With the switch "on", the needle will barely move no matter how one moves the tester, turn the switch "off", and proceed as usual.

-30-

Triode Amplifier (Continued from page 57)

inverse feedback for improving linearity, etc. In addition the circuit lends itself readily to direct coupling from The positive the preceding stage. voltage from the preceding plate circuit is more than offset by the high bias voltage developed across the cathode resistor, so that both tubes operate properly. The direct coupling permits extension of the low-frequency response to d.c. and eliminates coupling networks which can cause phase shift and low frequency attenuation. It will be noted from the schematic of Fig. 2 that in the "radio" channel there are only four coupling condensers in the entire amplifier, two on each side of the push-pull system. As both of these are of generous capacity, and the frequency response of the output transformer specified is no more than 1/2 db. down at 20 c.p.s., the low frequency response is very good.

The phase inverter is resistance-capacity coupled to a type 6SN7 tube connected in push-pull. A fairly high load resistor is used in order to increase the voltage output and the cathode resistor is not bypassed to take advantage of inverse feedback action in balancing the push-pull system. The stage is resistance-capacity coupled to a type 6SN7 tube used as a cathode follower driver.

The cathode follower is direct coupled to the push-pull power amplifier through an ingenious circuit. Each cathode is directly connected to a power stage grid. The cathode current flows through each side of a centertapped choke, which has low d.c. resistance so that no d.c. biasing volt-ages are developed. The choke has sufficient inductance so that its inductive reactance, which is the cathode follower load, is high, even at 20 cycles. The cathode currents from both sections of the cathode follower flow through R_{27} which develops the bias voltage for V_5 . The opposite end of R_{27} connects to the grid return of V_5 and the negative end of R_{28} , which is between the power transformer high-voltage center tap and ground, providing fixed bias for the power stage.

The power amplifier consists of a pair of 6A5 tubes coupled to the loudspeaker (see page 56) through the output transformer, T_2 . As the tubes are operated with fixed bias, the cathodes are grounded. The 6A5 is very similar to the 2A3 and 6B4 in characteristics and application, but is equipped with a cathode which is internally connected to the center point of the heater. The heater-cathode construction results in a considerably reduced hum level over the directly-heated tube types and, in addition, the internal connection provides a center tap ground for the heaters. No balancing arrangement for d.c. plate currents in the output tubes was provided because





with the Peerless hi-quality 18 watt TRIODE output amplifier kit. Heart of this kit is the Peerless S-240-Q output transformer, conclusively proved **superior** to all competition in **square wave** tests. Amplifier's response is flat from 20 to 20,000 cps for radio input. Phono input is equalized for variable reluctance pickup. Simple modification adapts amplifier for use with microphone. Kit contains all basic components and instructions for assembly.



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the unique circuit and the latitude possible in the output transformer made it unnecessary.

It will be noted that inverse feedback is included in the amplifier. Inverse feedback is usually thought of as being used with beam power amplifiers, but the circuit is not confined to beam tubes. The benefits of inverse feedback are just as real with triodes. Approximately 7 db. of feedback is used, the feedback loop including five stages and the output transformer. More than this amount of feedback could have been used, but it was found that this amount produced optimum results. It should be pointed out that the value of the feedback resistor, R_{29} , must be changed should the output transformer be strapped for other secondary impedances.

When the output transformer is strapped for a 16 ohm load, the internal impedance, looking back into the winding, is approximately 2 ohms. Thus the damping factor, the ratio of load impedance to internal impedance, is 8 and thus the amplifier will adequately damp a high efficiency loudspeaker.

The power supply is conventional, using two high capacity filter condensers and a resistor to provide humfree power. The resistor is used in the negative lead to provide, at the same time, grid bias for the power amplifier stage. (See schematic diagram, page 56.) The type 5V4 rectifier tube was used as it has a low internal resistance resulting in an increased d.c. plate supply voltage.

Another effect of primary concern to the user is hum. Assuming that the high voltage winding is properly center-tapped, hum can be produced by the stray magnetic fields from the power transformer. In some shell type designs, utilizing a single coil, the stray magnetic field may extend to audio transformers as far as three or four feet away, producing a hum that all the power supply filtering in the world will not eliminate. While the power transformer used in the A-100A Amplifier is of this shell type construction, it does have interesting features. The core operates normally at a flux density of 10,500 gausses and the cross sectional area of the core is uniform. The exception, of course, is at the joints where only one-half of the core is continuous. It is the concentration of flux at these points which contributes so greatly to the large external field normally encountered in small transformers of this type. Here though, the interleaving of the laminations is in a 2x2 arrangement so that the transverse flux path resistance is increased to the same order as that of the air gap between the single layer E and I pieces.

While this arrangement reduces to some extent the structure permeability and increases the reactive exciting current, it also produces a more uniform flux distribution in the core, particularly at the joints. The reduction of flux "hot spots" reduces the exter-

nal field. It also reduces the in-phase exciting current by an amount greater than the increase in reactive exciting current.

Therefore, two advantages result: (1) The external field is more uniform and less intense and (2) the over-all I^2R losses (in both core and copper) are reduced.

The completed amplifier performs exceptionally well when tested with square waves, the square waves showing at a glance the frequency response, phase distortion, and transient response. Fig. 5 shows traces of square waves passed through the entire amplifier. At 20 cycles the square wave shows some linear tilt which is caused by a slight phase shift. This tilt must not be confused with the exponential decay caused by lack of primary inductance often observed in inferior output transformers. The good high frequency response is confirmed by the fine square waves at funda-mental frequencies up to 20,000 cycles. There is no transient oscillation and only a little phase shift up to almost astronomical frequencies, as the square wave covers frequency and phase characteristics up to at least 20 times the fundamental. Square wave testing is becoming increasingly important as the performance of audio equipment is continually improved and has been widely publicized in connection with output transformer performance. -30-

HAMFEST

THE Starved Rock Radio Club of Uti-ca, Illinois, has planned its annual

Hamfest for Sunday, June 4th. The event will be held at Boy Scout Camp Ki-Shau-Wau like last year's hamfest. The usual arrangements are being made to assure an interesting time for all. Mobile operation will be stressed and those operators who keep a good log of their mobile work enroute to the meet will be rewarded for their trouble.

Registration is \$1.00 in advance or \$1.50 at the gate. Lunch will be available at the camp or those attending can bring their own lunches. Advance registration should be in the mail on or before May 25th to insure return of registration tags.

Registration requests should be sent to George E. Keith, Secretary, The Starved Rock Radio Club, Box 22A, Starved Rock Utica, Illinois.





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CHAPTER NOTES

Augusta—Camp Gordon

On March 3rd the chapter celebrated the 87th anniversary of the founding of the Signal Corps with a dinner-dance at the Camp Gordon Officers' Club. In keeping with the occasion the principal address, given by Lt. Col. Royal S. Copeland, was devoted to a review of the history of the Signal Corps and current developments.

Prior to the meeting, members availed themselves of the opportunity to review the exhibits and parade at Camp Gordon which marked the anniversary.

Baltimore

The Baltimore Chapter featured a talk on "Frequency Resources and National Policy" by Federal Communications Commissioner E. M. Webster at its February 23rd meeting. This was Commissioner Webster's second appearance before an AFCA chapter this year, having addressed the New York Chapter in January.

The meeting took place in the auditorium of the Pythian Building and was presided over by Chapter Chairman E. K. Jett of the *Baltimore Sunpapers* Station WMAR. He introduced Col. George P. Dixon, AFCA's new Executive Secretary, who spoke on "AFCA Present and Future."

Chicago

The Chicago Chapter's March meeting was held at the *Motorola Inc.* plant with Robert W. Galvin, executive vice-president of the company acting as host.

In the absence of Oliver Read, the Chapter's 1st vice-president, Dwight Brown, presided during the meeting. D. E. Noble's speech on "Motorola's Program for Decentralization of Facilities" was read for him as Mr. Noble was detained in Phoenix. Henry Magnuski spoke on the company's microwave equipment and John Doremus discussed "Recent Developments in Mobile Communications and Selective Signalling." A tour of the *Motorola* plant followed the talks.

Far East

Despite the handicaps of a widely scattered area and a transient membership, the Far East Chapter is carrying out a program which is receiving the enthusiastic support of its entire membership. Activities are conducted by the individual posts and are climaxed by an annual meeting of the entire chapter each spring.

The following is a round-up of post news:

Okinawa Post: New officers were elected at a meeting held at the Ryukus Command Signal Office on November 25th. Col. Joy T. Wrean, commanding officer of the 97th AAA Group, was unanimously elected president. Other officers elected were: 1st vicepresident—M/Sgt. Royle E. Morris; 2nd vice-president—Lt. Col. Clarence H. Lewis; Secretary—Capt. J. A. Beauregard; Treasurer—DAC Carl P. Dunkel.

Maj. J. W. Powers, outgoing chapter president, then introduced the Army and Air Force Military Amateur Radio System directors, and Capt. George H. Schmidt, of the 11th Signal Service Battalion, MARS director in Rycom, explained the proposed activities of his organization in the command.

The program closed with the film "The Battle of Okinawa."

On February 14th, members of the Okinawa Post toured the communication center of the Rycom Signal Section. After an explanation of visual screen teletype messages, the group sent and received teletype messages to and from the Tokyo Post of the Far East Chapter, AFCA.

Following this, visits were made to the ship-to-shore radio station, the center's terminal and relay section, the facilities control operations room and the Rycom telephone exchange.

Philippine Post: The Philippine Post was organized at Clark Air Force Base during the fall of 1949. Officers of the post are: Chairman—Maj. Forrest V. Diehl; 1st vice-chairman—Manuel A. Vargas; 2nd vice-chairman—1st Lt. Wyatt L. Law; Secretary—Capt. Bernard R. Grossman; Treasurer—SFC William A. Finney.





Tokyo Post: At a meeting of the Tokyo Post on January 10th, the following new officers and directors were unanimously elected: Chairman-Lt. Col. R. H. Mapes; 1st vice-chairman-Lt. Col. Jamie L. Wicker; 2nd vicechairman—C. B. Allsopp, vice-presi-dent of the International Tel & Tel Corp.; Secretary—Miss Nina Jo Cul-ver; Treasurer—Capt. D. V. Mayer.

Approximately 125 members and guests were present at the meeting. Entertainment featured showing of the film, "The Battle of Okinawa," and a social hour afterwards.

Yokohama Post: Grouped around the Headquarters Eighth Army at Yokohama, the Yokohama AFCA Post provides a fine medium for the acquaintance and association of the numerous communications personnel of the Army, Navy and Air Force in that area. Present officers of the post are: President-Lt. Col. Robert B. Tomlinson; 1st vice-president-Mr. Frank Hagen; 2nd vice-president — Capt. David M. Uhler; Secretary-Maj. Edgar DuBois.

Fort Monmouth

Although busy with arrangements for the national AFCA convention to be held there on May 13th, the Fort Monmouth Chapter took time out on February 9th to hold a chapter meeting at the Noncommissioned Officers' Club. In the absence of the president and vice-president on temporary duty. the meeting was conducted by Maj. J. D. Evans, Jr., senior member of the board of directors.

Short talks were given by S/FC Albert Cohen on the "Army Career Plan" and by M/Sgt. Pat Mulcahy on "Benefits of AFCA."

The main speaker was Capt. Henry E. Bernstein, USN, Director of the Armed Services Electro Standards Agency, who gave an interesting talk on the problems in connection with the operation and plans for the future of ASESA and the results being accomplished on a joint basis in standardizing the component parts of electronic equipment.

After the meeting, a program of entertainment was presented by the Special Services School of Fort Monmouth.

New York

Admiral Thomas C. Kinkaid, Commander of the Eastern Sea Frontier, was the guest speaker at the February 23rd meeting of the New York Chapter. The gathering was held at the Third Naval District's Officers' Club in the New York Naval Shipyard, Brooklyn, and was attended by 100 members and guests.

Admiral Kinkaid discussed the growth of communications in the Navy from the use of flag and blinker signals to modern radio and radar systems, and told a number of anecdotes about his experiences with Naval communications over a forty-year period. He described a few incidents in the Battle of Leyte Gulf, where he commanded the Pacific 7th Fleet, in which



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Sacramento

The Sacramento Chapter held its regular monthly meeting at the Sacramento Signal Depot on January 18th. Over 200 members and guests assembled to honor Brig. Gen. A. M. Shearer, Chief Procurement and Distribution Division, OCSigO, and former commanding officer of the Sacramento Signal Depot, and the conferees of the Depot Commanders' Conference (in progress January 16-20). The list of conferees included technicians, statisticians, engineers, supply officials, and key personnel from all Signal Depots and Signal Sections of General Depots within the United States, as well as top military personnel from the Office of the Chief Signal Officer in Washington. General Shearer was instrumental in the establishment of the Sacramento Chapter in November 1946 and served as its first president.

Chapter President Milton G. Mauer of the Pacific Tel & Tel presided and introduced the distinguished guests. Col. Harry L. Vitzthum, commanding officer of the Depot, welcomed the gathering and General Shearer delivered an impressive message on AFCA affairs and urged all in attendance to become members of the association. Murle C. Schreck, President of the Sacramento Chamber of Commerce, welcomed the visiting guests to Sacramento. Mr. Paul Shaad, Program Chairman, discussed plans for the next monthly meeting.

The main speakers for the evening were Major Joyce B. James and Major T. A. Pugh, respectively Executive Officer and Supply Office of the Alaska Communications System. The topics presented dealt with the operation and supply of the Alaska Communications System. The selected subjects were most timely and actual photographs were shown to emphasize the difficulties and problems encountered.

San Francisco

Members and prospective members of the San Francisco Chapter met on January 27th at the Don Lee Building to discuss plans for the reactivation of the chapter. The meeting was conducted by Harry E. Austin, local manager of RCA Communications, Inc., who is serving as president pending formal election of new officers.

In attendance were representatives of the local telephone company, pointto-point radio companies, broadcasting companies, and communications, electronic and photographic industries.

The meeting was addressed by Col. Lloyd C. Parsons, Signal Officer of the Sixth Army, Capt. C. C. Ray, Communications Officer of the 12th Naval District, and Col. G. B. Hoffman, Director of Communications for the 4th Air Force, each of whom presented the views of their service on the necessity for an organization such as AFCA and the active participation of the com-

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Stabilizer Cavity with bellows	gle wire dou- blet or coax. #1251-2	4- 9260 0c 926- 4- 926-	16 926-C13 -16 926-C13 -81 926B
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3 cm. "S" curve 6" long	sis, var cond, ind. 1 a m p. 60V 10c enclosed in Auto 6.9V	926A 926C	14 \$10.50
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DUPLEXER SECTION for 1124	Mfd Volt Price butt	eiver Fush- on 23-28	7" or 9"
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Navy type CABV-47AAN, with 4 in. slotted section 32.50	Condenser Kit, 43X- .05, 1-15 Mfd	3.1	.5.5 400 .50 1 600 .45
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polarized. 45 degrees	Plates for Control Box, 190-550, 6-9.1	150 180	1.5 1.5K .95 2 1.5K 1.05
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10 CM. WAVEGUIDE SWITCHING UNIT, switches 1 input to any of 3 outputs. Stand-	Relay, 24V Dpdt	57 58	1.5 6K 9.75 1.1 7K 3.39
plete with 115 vac or de arranged switching motor. Mfg. Raytheon. CRP 24AAS. New	123K7 Tube	62 67	1 7.5K 2.95 1 7.5K 12.95
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10 CM. OSC. PICK-UP LOOP, with male Home- dell output 2.00	PL154A—12 Png, Female	\$6.50	1 25K 83.95 4 50 .29 1 100 15
ball, with type "" or Sperry fitting	PL148A—3 Png, Female35 7027—18 Png, Male	Silver	2.5 100 .23
PHASE SHIFTER, 10 CM, WAVEGUIDE, WE	PL147-2 Png, Female	Feed Thru	2x,1 2ST .15 2x,1 3TT .15
MATCHING SLUGS 721A TR cavities, Heavy silver plated 200	Small3 Png Chassis, Female	500 470	2x1 4ST .16 2 2ST .20
10 cm. horn and rotating joint assembly. gold plated	Female	Button	.5 2TT .15 .5 2ST .15
TEST EQUIPMENT	7025—12 Png, Male	40 100 180 200	2x.25 3ST .21
Vide a means of rapid checking of crystal diodes 1N21, 1N21A, 1N21B, 1N23A.	.85 Mtd., 600V	185	2x.1 2TT .20 3x.1 3ST .25
1N23B. Operates on 1½ volt dry cell bat- tery. 3x6x7. New	7582—3 Mfd	\$8.50	.1 2TT .19 .5 1TT .19
10 cm, mfg. W. E. Complete with metor, in- terpolation chart, portable carrying case	2933-4X .8 Mfd	Write	2 2ST .23 2 2ST .26
head. New: Absorption type	7321—Variable Cond. 1.95 5032—Variable Cond. 1.25	for	.1 2BT .20
mission type, Type N Fittings, Veeder Root Micrometer dial, Gold Plated W/Calib, Chart.	5546-Rec. Choke Assy., RF.	of other	2x.1 3ST .27 2x.1 3BT .25
10CM ECHO BOX CABY 14ABA-1 of OBU-3, 2890 MC to 3170 MCS direct reading mi-	5634-Rec. Choke Assy., AF	Coils &	2x.1 2ST .26 3x.1 ST .25
9% to minus 9%. Type "N" input. Res- onance indicator meter. New and Comp. w/	0385—Relay Spst	Avail.	25 2TT .20 1 2ST .30
access. Box and 10 CM Directional Coupler. 350.00 3 CM. RECEIVER. SO-3. Complete with W.G. Mixer Assy. (723-A/B). Reg. Fil. Power Sun-	Available in start		2x1 3ST .35 2 2ST .40
ply 6 stages 1F (6AC7)	mitters, Receivers,	Mounts,	2BT .39 1000VDC
output. Includes UG28/U type "N" "T" junction and type "N" pickup probe. Mfg. cable, New	needs.	TE YOUR	MANY OTHERS
10 cm. cavity type wavemeters 6" deep. 61/2" in diameter. Coax. output. Sliver plated . ea. 64.50	A Heineman Bkrs. For	AC-DC	2 Receiver w/Tubes omplete 1.ess
115v D.C. Tuning motor sub sig 118Aea. 47.50 W.E. 1 138. Signal generator, 2700 to 2900 Mc. range. Liebtbuuga tota	B	A m - D 1, 12,	d 500 W C:11-
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munications, electronic, and photographic industry.

Washington

Major General Spencer B. Akin, Chief Signal Officer of the Army, was the guest speaker at the Washington Chapter's luncheon meeting on March 7th at Fort Myer, Va., which commemorated the 87th anniversary of the founding of the Signal Corps.

During its 87 years of existence, and particularly during the last war and the postwar period, the Army Signal Corps has accomplished notable progress in communications operations, research and development, and procurement, distribution and training, because it is a member of a "team composed of science, industry and armed services," General Akin declared.

After the luncheon, which was attended by 150 members, the group toured the Signal Corps station WAR, and the mobile communications center at Fort Myer and the Army Communications Center at the Pentagon. -30-

TRANSMISSION LINE

A revolutionary telephone and television transmission line has been developed by the U. S. Army Signal Corps and dubbed the "G-String" in honor of its inventor, Dr. Georg Goubau.

The line, which is simple, highly efficient, and costs little to make, promises to open up wholly new possibilities in microwaves and home TV. It is a single wire with special insulation and funnel-shaped terminals.

The Signal Corps expects that the new line may eventually replace the more costly coaxial cable for transmission of television programs. -30-

SIGNAL CORPS PROGRAM

WHAT promises to be the most elaborate communications and eleetronics program in Signal Corps history will be presented at Fort Monmouth, New Jersey, on Saturday, May 13th under the sponsorship of the Fort Monmouth Chapter, Armed Forces Communications Association and the Signal Corps.

The Fort Monmouth program is scheduled for the day following the annual National AFCA convention at the Commodore Hotel, New York, under the auspices of the New York Chapter.

The scheduled all-day program at Fort Monmouth will be highlighted by the presence of leaders in the communications and electronics fields and high-ranking officers from all branches of the Armed Services.

Convention delegates will be transported from convention headquarters in New York to Fort Monmouth in chartered buses. The program will feature elaborate displays from the Signal Corps Engineering Laboratories, the Armed Services Electro-Standards Agency, and The Signal School of Fort Monmouth; parachute drops by the 82nd Airborne Division; wire laying demonstrations by helicopter and bazooka; and a combat communication problem in which 82nd Airborne troops will figure, weather permitting. -30-



PS PROGRAM SERIES "B"

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Electronic Metronome (Continued from page 45)

or when room noise is high, a larger speaker may be used.

The power supply consists of the rectifier section of the tube and two filters. The first filter is comprised of resistor R_a and the dual 40 μ fd. condenser, $C_{4-}C_{5-}$ Voltage supplied to the oscillator is regulated by the OC3/ VR105 tube. The second filter, consisting of resistor R_2 and condenser C_{1} , delivers plate and screen voltage to the pentode section of the 117N7-GT. This type of power supply enables operation on either a.c. or d.c.

Construction

Fig. 1 shows the external appearance of the complete metronome removed from its cabinet. Fig. 3 is an under-chassis view, showing placement of components. In Fig. 1, the tubes (left to right) are neon lamp, 117N7-GT, and OC3/VR105. The lefthand shaft is the volume control (R_3), the right-hand shaft the speed (frequency) control (R_4).

The instrument is built on a boxtype metal chassis; $6\frac{14}{7}$ long, $3\frac{1}{2}$ wide, and 2" high. The entire device can be housed in a small radio or intercom cabinet.

The wiring is simple and straightforward. Keep all leads as short as possible. For safety to the operator, do not connect any part of the circuit to the chassis.

Adjustment

If the wiring is done correctly, the speaker will begin clicking off beats as soon as the 117N7-GT has come to operating temperature and the VR105 has started to glow. The lowest setting of the volume control will reduce the beats to the below-audible level. The beat speed will be found to vary smoothly as rheostat R_4 is rotated.

A dial plate, graduated directly in beats per minute or per second should be attached to rheostat R_i . This dial may be calibrated by means of a stop watch or any other watch having either a sweep second hand or an easilyread conventional second hand.

Fig. 3. Under chassis view of the metronome.



May, 1950



TELEVISION SCOPE

SUPERIORITY AT A GLANCE!

The vertical response of this economy TV scope is usable to 5000 kc, not 50 kc. Response is flat to 750 kc, down 3 db at 1000 kc. Amplifier supplies a valtage gain of 20 at 5000 kc.

AR-3

DETROIT

Check this necessary feature before you buy any scope for TV use.

The R.S.E., AR-3 Scope has been built by Ross Armstrong to our rigid specifications. It's a complete unit that embodies standard horizontal amplifier and sweep circuits with normal sensitivity.

The case is 8" high x 5" wide x 14" long, attractively finished in "hammered" opalescent blue enamel. Operates on standard 110 volts —60 cycles—40 watts. Tubes, 3BPI-6AC7-6SJ7-6X5-5Y3-884. Instructions included. Complete specifications upon request. Satisfaction or your money back. AVAILABLE TO JOBBERS AVAILABLE TO JOBBERS F. O. B.

IN QUANTITY







It is also possible that the calibration might be made with the aid of a variable-speed recorder (wire, tape, or disc). Place the metronome speaker before the recorder microphone and record a full minute of operation at a given speed. Then play back the recording at a slower speed and count the beats per minute (sum total of the recording). -30

Built-in Antenna

(Continued from page 59)

able matching device the maximum amount of signal can be transferred into the front end. As can be seen from Fig. 1, the Philco built-in antenna is shaped in a tapered form with the narrowest part in the center. This form allows for a relatively broadband response over both the low and high bands. The antenna itself is made of two sections of aluminum foil 0.005" thick, which are attached to the tunable matching device. The foil is mounted to the underside of the top of the cabinet, and its ends are folded over. This folding has the effect of adding capacity and thus lowering the natural resonant frequency of the dipole. This increases the electrical length of the antenna.

Tuning this system consists simply of tuning a variable condenser connected across the terminals of a hairpin coil which has two shorter loops attached to it at the antenna end. A long wooden rod is attached to this condenser and projects through the cabinet front so that it can be adjusted.

To understand just how this system is tuned it is necessary to fall back on resonant circuit theory. The dipole may be represented as a series resonant circuit. This is shown simply in Fig. 2. As is known from resonant circuit theory, a series resonant circuit will look inductive below resonance and capacitive above resonance. The tunable matching network may be represented as a parallel resonant circuit. This is also shown in Fig. 2. It can be seen that, above resonance, the circuit is inductive; below resonance, it is capacitive. If we draw a curve for this effect and plot reactance $(X_L \text{ and } X_C)$. against the frequency, we obtain Fig.

Fig. 4. Position of Philco's built-in antenna tuning knob in relation to the twelve assigned video channels.





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Fig. 5. Simplified sketch of the coils used for tuning the Philco electronic built-in antenna. A and B point to the long "hairpin'' coil to which is connected the 300 ohm line and the two shorter loops (C).

3. To make the entire antenna system resonant and thus effect maximum transfer of energy into the receiver, it is necessary to cancel out the reactance of the antenna with a reactance of equal amount but of opposite sign. Thus by tuning the parallel resonant matching system, the relative position of the curve for reactance for the parallel resonant circuit is shifted in respect to the series resonant curve, so that the reactance of the antenna to load is matched throughout the band. A hairpin loop containing an inherent amount of inductance resonates with the tuning condenser to form the parallel resonant circuit.

The built-in antenna system is tuned for each channel to an approximate impedance of 1500 ohms, with the actual variation being from approximately 1200 to 1500 ohms. The 300 ohm impedance of the line is matched by tapping on to the proper feed point on the long loop. See Fig. 5.

Essentially the advantage of such a system over an untuned type consists of a higher signal input to the receiver. Besides this, tuning the system will eliminate standing waves and prevent their effect on the picture. It has been reported that it will also provide a means for "tuning out" interference of various kinds, such as adjacent receiver radiation and some types of FM interference.

The result is better reception generally with this type of tunable or "electronic" antenna system. -30-

CORONA AND ARCING By MATTHEW MANDL

ELIMINATING corona or arcing in television high voltage power sup-plies often is a time-consuming process because it may require complete re-moval of the high voltage shield casing and several components before the place where the leak occurs is accessible. In many cases considerable time can be saved by using celluloid or mica strips because these can often be inserted between points of arcing more easily than moving components farther apart in an effort to reduce high voltage leak-

age. When the corona or arcing in the high voltage system has been located, only one panel section of the high voltage shield compartment need be removed and the strip of mica can usually be inserted by means of thin-nose pliers. The insulating strips should be cut so that they fit snugly between the parts where the arcing occurs. Thick celluloid or mica is preferred and the latter can be salvaged from the heating elements of old toasters or electric irons. -30-

May, 1950



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Antenna Equipment RC-173 is an ultra-high-frequency and low frequency antenna system which provides either directional or non-directional ultra-high-frequency anten-nas for use with several radio receivers simultaneously. It consists of an insulated 30 foot, 3 section tubular steel mast and three tiers of dipole assemblies, two in each tier at right angles to each other. Each dipole assembly covers a specific frequency band within the range of 27-145 mega-vales. a spec cycles. The When cies. When erected, the mast will withstand high wind velocities and adverse weather conditions.

Complete system consists of 30 foot 21/2" OD Mast, in three sections, no taper. Elements: Steel, 10 each, Tier #1: 23"; Tier #2: 41"; Tier #3: 71", w/assemblies.

Coaxial Cable RG-22U 95 ohm; 6 of 50 ft. and 2 of 40 ft. lengths; RG-11U 75 ohm; 1 of 21 ft. length w/fittings and terminal box. Also Guys, Gin Pole, Stakes, etc. as illustrated. Shipping Weight: 400 \$4950 \$49<u>50</u> tc. as illustrated. Sing bs. Price, only lhs

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NEW TV PRODUCTS _____on the Ma ket

19-INCH CONSOLE

Among the television receivers unveiled by *Motorola*, *Inc.* recently is the



Model 12K2 console which retails in the moderate price class.

Featuring a furniture-styled cabinet and simplified controls, this 22'' wide, 36'' high, and 18%'' deep set is available in either mahogany or limed oak finishes.

The receiver has the new "flywheel" a.f.c. circuit which keeps the picture clear and steady despite noise and changes in signal strength, according to the manufacturer. A built-in antenna and target tuning are also features of this receiver.

The sets are being manufactured at the Company's plant at 4545 Augusta Boulevard in Chicago, and further information may be obtained from this address.

SIGNAL GENERATOR

A miniature TV-FM signal generator for the service technician has been put on the market by *Oak Ridge Products*, 239 East 127th Street, New York 35, N. Y.

The new model "103" is said to pinpoint any signal failure from antenna to cathode-ray tube or speaker in less than two minutes. Incorporating three



separate tuning bands and modulation output and attenuator for TV and FM, the unit generates a signal to perform as a complete r.f., oscillator, and mixer tester, video i.f. tester, audio i.f. tester, and as a checker for the video and audio second detector, the video and audio amplifiers, sound traps, adjacent picture traps, and as a marker generator, antenna orientation and sensitivity tester.

TV ACCESSORIES

The LaPointe Plascomold Corporation of Unionville, Conn., has added two new items to its line of television accessories.

Production is under way on a lightning arrester for use with the fourwire control cable employed with antenna rotators. Designated the Model RW 204 arrester, the new unit is designed to serve dual purposes in that it may also be used for standard 300 ohm ribbon transmission line. It is supplied in a low-loss polystyrene case which will not absorb moisture nor deteriorate from weather. Pin point contacts eliminate the need for wire stripping.

The second item recently introduced is a three-way antenna switch for multiple antenna installations. The new unit allows the use of a separate transmission line for each antenna in



such installations and provides maximum performance from each antenna. By a simple turning of the knob the viewer can change over instantly from one antenna to another. The switch is also suitable for television dealers when demonstrating more than one receiver from a single antenna.

MATCHING TRANSFORMER

To prevent mismatch between a folded dipole antenna and the coaxial lead and to prevent loss of signal, *American Phenolic Corporation* of Chicago 50, Illinois, has recently developed a new matching transformer.

The new unit employs Amphenol four-conductor twin-lead specially connected to provide a transformation in impedance from 300 ohms down to 75 ohms. It is particularly useful in matching 300 ohm transmission line to 72 ohm or 75 ohm input television receivers or providing an efficient match between any 300 ohm antenna and a coaxial 75 ohm line such as might be used in an area of high noise level.

The terminals on the transformer are spade type lugs at each end of the

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transformer section which should be operated fully open or in a very loose coil. Tight coiling is unnecessary, according to the company.

HIGH-VOLTAGE PROBES

Two new high-voltage probes, designed for use with popular lowcurrent voltmeters to provide safe, convenient and accurate means for measuring high voltages in high-im-



pedance circuits such as those found in television receivers and broadcast transmitters, have been announced by the Tube Department of Radio Corporation of America.

The two probes, Types WG-289 and WG-290, are identical except for their connectors. The WG-289 is designed for use with electronic voltmeters and is equipped with a microphone-type connector. The WG-290, for use with non-electronic voltmeters, has phonetype connectors. To adapt the probes to various popularly-used makes of voltmeters, five types of multiplier resistors are available separately.

The probes are being handled by RCA test equipment distributors.

BOOSTER KIT

Tech-Master Products Company, 443-445 Broadway, New York 13, New York, is marketing a television booster kit, the Model 3375.

The unit, which is easily assembled, features fixed grid and variable plate tuning, separate coils for low and high channels, a tunable high channel input coil, pre-stamped chassis for ease in mounting parts, wood cabinet, simplified step-by-step wiring instructions, and adaptability for either 300 or 72 ohm lines.

The booster kit comes complete with the 6AK5 and pre-aligned coils. A data sheet on the kit is available.

ATTIC BEAM

Telrex, Inc. of Neptune Highway, Asbury, Park, N. J., has just introduced a new television receiving antenna expressly designed for indoor installation. Known as the Attic "V" Beam, the new unit may be easily and quickly installed. It requires no masts, towers, or guy wires and is free from all hazards.

The entire assembly folds into a compact package for easy handling
and placement and then opens up completely assembled into a full two-bay Conical "V" Beam array, complete with transmission line. The array is



extremely light and entirely self-supporting. It may be suspended or rested on flooring, beams, or rafters.

DISTRIBUTION UNIT

Lynmar Engineers of 1721 Delancey Street, Philadelphia 3, has developed a new distribution system for multiple television receiver operation from a single antenna.

Designed for apartment houses, showrooms, and service shops, the new unit covers both high and low bands without switching. Feedback from one television receiver to another through the equipment is said to be negligible.

The standard unit for feeding six television receivers is composed of three coupling units, each of which feeds two television sets, a power supply, and a mounting cabinet.

More than six receivers may be operated from the same antenna by connecting additional assemblies in series. The equipment has been standardized for 52 ohm antenna input coaxial transmission line. An impedance matching transformer is available for use with installations where a 300 ohm input line is used.

A flyer on the #PD-16 is available on request.

TV-FM AMPLIFIER

Sonic Industries, Inc. of 221 West 17th Street, New York is producing a new "Super Sonic" TV-FM amplifier, the Model IT4.

Designed to provide high gain preselection for any television or fre-



quency modulation receiver with adequate bandwidth to pass all desired modulating elements and provide adequate selectivity to reject unwanted off-carrier signals and noise, the new unit has one continuous tuning control and one off-on switch, without any verniers or Hi-Lo switching.



Why shop around?

It's easy to get the right equipment from RCA's full line of matched sound products

Buy your sound products the easy way. The wide variety of RCA Sound Products simplifies your problem of finding the right equipment for your sound jobs.

Every item in RCA's extensive sound line is electronically engineered with its own special characteristics to give top performance, dependable long-life service with quality appearance. Not only is the RCA Sound Products line built right, it is also priced right to enable you to build a steady volume of profitable business.

No other manufacturer offers so extensive a line of "matched" sound products from the smallest "ballyhoo" system to the largest systems for industrial, educational, institutional, church, hotel, hospital or commercial users.





SOUND PRODUCTS RADIO CORPORATION OF AMERICA ENGINEERING PRODUCTS DEPARTMENT. CAMDEN. N.J.





ELECTRO PRODUCTS LABORATORIES, INC., 4509 N. Ravenswood Ave., Chicago 40, III.

May, 1950



Gives Tube position and function in hundreds of important TV receiver models, made by 56 manufacturers.

FIND THE TROUBLE AND REPLACE TUBES WITHOUT REMOVING CHASSIS

levisio

Nothing like it! The only book that shows the position and function of tubes in hundreds of TV receivers. Often an operational check in the customer's home . . . looking at the picture tube and listening to the sound . . . can give you a clue to the trouble. Many times only a tube failure is responsible. TGL-1 makes trouble diagnosis and tube replacement quick and simple, in most cases without re-moving the chassis! Each model has its own clear, accurate diagram. Book fully indexed for quick reference. Over 200 pages, handy pocket size, $5\frac{1}{2} \ge 8\frac{1}{2}$ ". Get two copies . . . one for outside calls and one for your bench. Pays for itself on the first job!



STRINGING GUIDE" NEW!

SECOND VOLUME Covers models from 1947 to October 1949

Over 45,000 servicemen bought the first volume of this invaluable book! New second volume includes 511 different dial cord stringvolume includes 511 different dial cord string-ing diagrams used in almost 1000 receivers produced from 1947 to October, 1949 (all new data continuing from where the first volume left off). There's only one right way to string a dial cord... and here's the only book that shows you how. Saves time—saves effort. Handy pocket size Order copies for some teel Handy pocket size. Order copies for your tool kit and work bench today. \$ 00

ORDER DC-2 Only
HOWARD W. SAMS & CO., INC.
Order from your Parts Jobber today, or write direct to HOWARD W. SAMS & CO., INC., 2201 East 46th Street, Indianapolis 5, Ind.
My (check) (money order) for \$ enclosed. Send the following books: TGL-1 "TV Tube Location Guide" \$1.50 DC-2 "Dial Cord Stringing Guide" \$1.00
Name

Balanced input and output circuits provide for minimum noise pickup. Versatility in application has been provided by both 72 ohm and 300 ohm input and output matching impedances. An isolating power transformer is used to keep input and output circuits above the power line circuit.

ROOF BRACKET

A low-priced roof bracket which may be adjusted to hold television antennas on the peak of a roof, on the



slope, on the flat, or on the side, is currently being offered to the trade by TV Accessory Engineering Co., of 2105 Lagrange St., Toledo 8, Ohio.

The bracket comes prepared for use as a roof saddle but may be bent into any necessary shape to permit its use in other locations.

RAYTHEON'S "ROVER"

As one of the company's "Silver Anniversary" line of television receivers the Belmont Radio Division of Raytheon Manufacturing Company has introduced "The Rover," a compact table model video set.

Housed in a maroon tooled leather-



ette cabinet, "The Rover" (Model M-1106) carries a 12¹/₂-inch, glare-free, black picture tube which provides a full 110 inches of viewing surface. The set uses a built-in antenna.

The cabinet measures 19" wide, 18" high, and 19" deep.

TV ISOTAP TRANSFORMER

In an effort to promote safety and efficiency in the testing and servicing of television and radio receivers and other electronic and electrical equipment, the Tube Department of Radio



ADDRESS

CITY STATE

Corporation of America has announced the availability of the "Isotap," a unique test instrument.

Combining in one compact unit an autotransformer and an isolation



transformer, the new television "Isotap" is designed to speed up television receiver service, minimize shock hazards, cut down service returns, and prevent damage to costly shop test equipment.

TV FILTER

Norval H. Copple, of 3701 North Broad Street, Philadelphia 40, Pa., is handling the distribution of a new television filter which has been tradenamed "Nutro-Lite."

An optically correct filter, the new unit was developed by Kern-Moss Op-

tical Co. According to the company, the filter rests the eyes while viewing and eliminates all chromatic distortion and reflections to give a clear, sharp, flickerless, and easy-on-the-eyes picture. The filter is said to work equally well on both the conventional and black tubes with no additional brightness being required. The "Nutro-Lite" filter is available

The "Nutro-Lite" filter is available in two types. A single clip-on lens which is placed against the outside of the safety glass and held there by inserting small metal clips between safety and frame and the type whereby the filter lens is incorporated into the laminated safety glass which replaces the plain safety glass supplied by the set manufacturers and is known as a custom-made installation.

MATCHING TRANSFORMER

An impedance matching transformer which can match a 70-95 ohm coax line to a 300 ohm input has been developed by *Lynmar Engineers* of 1721 Delancey St., Philadelphia.

The unit may be used to match a 300 ohm antenna to a receiver made for 70-95 ohm coax. In receivers with both coax and 300 ohm twin-lead connections, improvement in reception is possible when the input is connected through this transformer. $-\overline{30}-$

ZENITH TO TEST PHONEVISION IN CHICAGO

HAVING received FCC authorization to conduct a series of Phonevision tests in the Chicago area, Zenith Radio Corporation is currently setting the stage for the 90 day test.

A group of 300 families, representing a cross-section of the community and selected by the National Opinion Research Center of the University of Chicago, have been named and currently television receivers equipped with Phonevision and the necessary telephone facilities are being installed in these homes.

As to the operation of this new type of television, the company explains that in this system the Phonevisiontelevision transmitter codes, or scrambles, the television picture. The receiver, with the key signal provided over a telephone wire, decodes or unscrambles it. A Phonevision receiver is a standard television receiver with a decoding unit built in or added to the original receiver. It will receive any of the desired "free" programs that are within range in addition to the subscription programs if it is set up to receive the key signal over the telephone wire.

When a television broadcaster wishes to transmit a subscription program and prevent its unauthorized reception he merely throws a switch which connects a special unit into the circuits of his transmitter. From that moment on, the subscription program is broadcast with certain variations which produce a jumpy, uncertain image shifting back and forth across the television screen at a rapid but entirely random rate. In this form the picture has no entertainment value and cannot be watched for more than a few seconds without eyestrain.

To make the program available to those set owners who wish to pay for it, the transmitter generates a key or decoding signal which passes over the telephone wire to the telephone exchange. From the exchange, this signal can be sent over the regular telephone wire that serves any home and delivered to the television receiver of any subscriber who has asked for the service. In the television set the key signal actuates a decoder which automatically corrects the broadcast signal sent to the television set over the air. The random shifting of the image on the screen is stopped and the picture becomes clear.

If commercial Phonevision is established it will operate as follows. The set owner consults his daily newspaper or program booklet to select the subscription program scheduled for that day. Next he contacts the Phonevision operator to request the key signal for the program of his choice. The operator then makes a record of his telephone number and makes the connections which will send the key signal over his line at program time. Shortly before the scheduled hour the set owner tunes his receiver to the proper channel. When the subscription program begins, it is received in clear form. At the end of the month the set owner is billed for the fees for each picture he has ordered. According to Zenith, Phonevision will

According to Zenith, Phonevision will not interfere with the normal use of the telephone. A filter arrangement added to the telephone wire stops the TV signals from getting into the telephone, prevents the key signal from reaching the telephone, and blocks telephone impulses for disturbing the television sets. In cases of party line telephone service, different frequencies will be used for transmitting the key signals to the two different subscribers on the party line.

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your FREE Folder now. Examine it-use it





in better television receivers!

The new QUAM FOCALIZER UNIT, utilizing Alnico permanent magnets instead of the more expensive and less stable wire wound construction, not only provides better, sharper focus of the television image, but is unaffected by operating temperature and voltage fluctuations. With a QUAM Focalizer Unit in the set it is unnecessary to refocus the image

AND

maintenance.

cal examination.

investigation.

year.

higher.

because the coil has heated, increasing resistance and weaken-ing the magnetic field. Since there is no wiring, voltage fluctuations have no effect on the operation of the QUAM Focalizer.

Focusing is done upon installation of the set by a simple Focalizing Unit in the better Television Receivers. THE QUAM ION TRAP

The improved and simplified construction of the QUAM ION TRAP, also utilizing permanent magnets, makes it preferred equipment for better television receivers.

QUAM-NICHOLS CO. 33rd Place and Cottage Grove CHICAGO, ILLINOIS Makers of QUAM Adjust-A-Cone SPEAKERS



Extra Income

(Continued from page 47)

The recording turntable, cutter head, and playback table can be inexpensive "mail order" types; occasionally a bargain can be found in the "swap' sections of catalogs and trade magazines. More professional-type equipment can be added later as your recording business grows.

Improving Receivers

Have you ever serviced a receiver, completely aligned all stages, replaced all defective parts, and still were not satisfied with its perform-ance or sensitivity? This is more common in receivers several years old, particularly the a.c.-d.c. type. (In many cases this is a result of poor design and not much can be done.) Sometimes, however, a simple circuit change or part replacement will help. Here are a few things you can try: (1) If the set is worth the cost, replace the i.f. transformers with iron-core units. (2) The "Q" of the oscillator and r.f. coils may be low, too, due to absorption of moisture, high coil form loss, etc. Replace with improved types recommended by the manufacturer. (3) Decrease the value of the a.v.c. return resistor. (Try paralleling a 1/2 megohm resistor from the r.f. end of the a.v.c. filter resistor to ground.) This reduces attack time and raises the gain. (Avoid this change if strong local stations tend to overload or distort.) (4) Performance in some sets can be improved by raising plate and screen voltages. (Keep within tube ratings; be sure bias voltages are correct.) (5) In some a.c.-d.c. sets, there is excessive interaction between the chassis and the unshielded loop antenna. Move the antenna farther away if possible or insert a thin sheet of copper between loop and chassis, and ground the sheet to chassis. (This increases the "Q" of the loop.) (6) On receivers having no trimmer on the detector section of the tuning condenser, try adding one. (Use a surplus unit or a trimmer from a discarded i.f. transformer.) (7) In extreme cases, try leaving the tube shield off in the converter or one of the i.f. stages. This amounts to increased interstage coupling and controlled regeneration up to a certain point. If receiver oscil-lates, detune one of the i.f. stages slightly or discard the idea. (8) Some owners of well-designed receivers prefer hi-fi local reception to sharp DX tuning. Try "flat-topping" the i.f. stages to give a 10 kc. bandwidth; note the improvement in tone quality and reduction in noise. (A scope will be needed for symmetrical \pm 5 kc. alignment.)

Battery Set Conversion

Converting battery sets to a.c. operation is a relatively simple operation. The simplest method is to install a power supply or converter

whose "A" and "B" voltages correspond to those of the battery set. Commercially-built battery eliminators are available for less than it would cost to construct a similar supply in the shop. Straight battery-operated sets are relatively few in number and can usually be traded in on a new a.c. receiver more economically than converted to a.c. operation.

Rectifier Applications

Much has been published about selenium-cell replacement of rectifier tubes. Here are a few additional uses for selenium cells which may be helpful around your shop: (1) As a.c.-d.c. radio power supplies, (2) battery chargers, (3) battery eliminators (automobile radio testers or demonstrators), (4) coin-operated machines (quite a few service technicians work on these units), (5) electrical musical instruments (guitar amplifiers, etc.), (6) exciter lamps (theaters, homemovies), (7) farm radio converters, (8) field supply for speakers, small d.c. motors, etc., (9) filament supply (battery and portable radios, a.c. amplifiers, etc., where a.c. hum component must be kept very low), (10) hearing aid battery chargers, (11) high-voltage power supplies for p.a., TV testing, condenser breakdown tester, etc., (12) model trains and other electrically-operated toys, (13) motion-picture projectors (carbon arc lamps, etc.), (14) polarized relays (battery chargers, current-limiting devices), (15) rectifier-type instruments (test equipment, a.c. meters), (16) relay operations, (17) scientific instru-ments (ore locators, Geiger counters, seismographs), (18) therapeutic equipment (heat lamps, machines), (19) Time clocks (in factories, business offices, plants), (20) Tone generator (in p.a. systems used by factories, schools, large offices, etc.), (21) vibrator-type supplies (rectifiers in auto sets, testers, etc.), (22) voltage regulators (a.c. and d.c. circuits), (23) spark quenchers on motors, d.c. commutators, etc. (for eliminating arc-quenching resistors and condensers; also protecting and minimizing arcing across d.c. contactors. Acts as interference filter).

Custom Building

It is nice to think of the profit in one deluxe FM-TV custom-built installation. It is also nice to remember that contracts for such installations are often the result of having done a good job on lesser installations, radios, and even appliances.

Possibilities in your community are better right now for custom-built systems than ever before. More new homes and apartments are being constructed every day, and a large percentage of these will definitely be prospects for some type of built-in or custom-built system. (Custom building need not be overly expensive to the customer, nor does it necessarily apply to new equipment alone. The customer's old radio, record-changer, etc., may be used if he wishes. Here



-ARMY WAREHOUSE CLEARANCE -
HS-33 Headsets. We got a beautiful lot of these 600-ohm (red plug) sets. Two piles: HS-33 so clean that we could, if we wished, wrap into cartons and call them new (and they prob- ably are)
HS-33 with discolored headbands
marker pips at 2, 20, and 100 KC. Two parallel horizontal sweeps; time the differences between sigs, between half power points on pass band curves, and many other scope uses. Double-deck tube shelves, lots of room for any circuit you can dream up. For
example, make start-stop sweep circuit to start both xtal and sweep, and one cycle of your input sig, will always be stationary. Contains 27 tubes, 22 pots, loads of switches, condensers, transformer and small parts Used With schematic Lowest
prices ever offered. Exc. condition 524.95 Fair used, may be short the xtal and 517.50 Few tubes
indicator described above. Easily converted to 60 cy. operation for use as a 160 meter, 4 chan, recr. or as a high voltage power supply for oscilloscope. Used. With schematic and conversion \$5,95
ATTENTION AIRLINE SHOPS! Antenna AT-4/ARN-1, replacement for radio altime- ter. With undamaged porcelain ring and no \$2.95
With damaged ring and/or rattling bead, can mibalize for housing and SO fitting
pack, freq. range 2500 to 6700 KC. Has 0-1 amp. RF meter in face. Controlled at front panel. 2 xtal channels. 6L6 output. Only 15.4 lbs. \$17.50 Used
BC-375 100 watt xmtr unit, 200 KC to 12.5 MC when used with appropriate tuning units. With sche- matic showing complete parts description for use with dynamotor PE-73 and also with AC power sup- ply conversion instructions. Xmtr only, used, snecial
Army warehouse clearance price, while \$12.95 Tuning Units for BC-375. Specify freq. coverage desired. Excellent condition. With sche-
matic showing parts description
is and modify a second to cover lower TV channels. Simple DC power, needs only 2 batteries (1.5 v and 45 v.). Precision vernier dial. Hand calibrated tun- ing charts. Diode-triode tube, 0-500 DC micro-am- meter. Pickup from tuned resonant cavity is detected
by flode. Negative DC is applied to triode grid, and dips the plate current. Complete, with plug-in an- tenna, crackle aluminum carrying case. $(12)_2'' \times 83_4'' \times$ 6'Jz'') with leather handle. Schematic in 512.95 side. In original wood storage case.
HERE'S A HONEYI Range Calibrator I-233. Built-in 115 v., 60 cy. power supply, Xtal, 186.30 KC, generates accurate
Ac, also counted down sync pulse. With test cords. Schematic inside case cover. With xtals and tubes: 2-6SN7GT. 1-6L6G. 1-6V6GT. 1-6SJ7, 1-5Y3GT. Externally dirty, clean as hound's tooth inside. Lots
Pulse attenuation. Voltage Divider TS-89/AP. Use your low range voltmeter or your scope to measure of the state of the sta
ranges, 10:1 or 100:1. Accurate RC circuits pre- serve exact waveform, flat from 150 cy. to 5 mc. In small, neat case with HV insulators and coax cables and plugs for connection to scope and CA OF
Remote Scope. Indicator ID-41/APQ-13. Radar PP1. 5PP7 in stepped cylindrical container with beautiful set of selsyn-driven coils. With selsyn. Other con- trols: Vert. cent. Hor. cont. Respire Respirement
Scale Illumination. Plastic scale can be hand ro- tated against fixed compass calibration. 6AK5 tube and input circuit in back end. Ranges are 4, 10, 20, 50, and 100 nautical miles. With cords attached
PP-4/APQ-2 Rectifier power supply. Consists of two thermionic rectifiers with separate filter circuits which supply independent with separate filter circuits
ten times selling price: 4-5R4GY tubes worth \$1.25 ea., 2-1mfd, 1500 VDC or 1000 VAC condensers worth \$1.50 ea., 24 mfd, 1000 VDC or 660 VAC con- densers worth \$1.50 ea., 400-2600 cy. transformers; chokes bloader resistors world sarriers;
Net wt. 37 lbs
Here's a recvr. that you can do wonders with! High-Q 90 cv. only and 150 cy. only filters. Make inter- modulation analyzer for testing quality of audio am- plifiers by putting 60 cy. and 90 cv. (buzzer hash thru 90 cv. filter) to input. Output them 150 cr.
filter, divided by input, gives percentage of modula- tion distortion. Contains dual bridge instrument rec- tifiers. Also best AVC system yet devised, flat from 5 to 100.000 uv. uses 100 VDC RF power supply for AVC blas. lift out intact to wire into commende
RF is 108.3 to 110.3 as selected by 6 relays in local oscillator. 18th harmonic beats with RF to give 6.9 MC IF. Lots of IF amplification, excellent for FM Broadcast use. Set has 3-717A tubes, use as direct replacement for your 6877
gain. Also 2-12SF7, 1-12SQ7, 1-12A, 1-12AH7, 2-12SR7. With schematic, and conversion 53.95 done to home FM recvr. Excellent condition 53.95 BC-347-C Booster Amplifor. excellent condition
input amplifier to boost modulation, excellent as infer- mike or phono preamplifier. Compact, only 534,"x4"x 215,": contains dual-triode power tube 678G and two hi-fi ouncer transformers, one for mike to push-rull grids. One for push-pull output With approximation
Requires external power source. Original use was for interphone amplifier. Good used, a special 69c 1-82-A Autosyn angular position indicator. Navica.
tor's ADF indicator with rotary reference compass scale. Use for beam-position indicator with our \$3.79 loop, motor, and autosyn-xmtr. Special Army warehouse clearance price, while they \$7 95
Desk Mike T-48. Desk telephone xmtr. Weighted bottom. Excellent ham shack mike. Long \$1.95 rubber covered cord terminating in PI-68 plus \$1.95
Precision Resistor Special. Control Box, Angle Se- lector. C-230/APQ-7. Cannibalize for: A nice black- wrinkle aluminum case 3"x4"x5"; a 9-throw simele pole rotary selector switch with standard pointer knob; a SPST spring-loaded snap-action push switch-
and precision resistors. Some models have 21 matched gold-band 1/2 watt resistors, others have eleven 790 1/4 % resistors. Soecial at only
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1200 S. Alvarado St. Los Angeles 6, Cal. All Prices F.O.B. L. A. Calif. Buyers add Sales Tax. SEND FOR OUR LATEST CATALOGUE!

again the demonstrator idea in Fig. 1 will help the customer decide which course to choose.) There are many good FM-AM tuners, amplifiers, speakers, and record-changers on the market, and a combination of any of them will work well.

Placement of each unit depends on the customer's wishes and the arrangement of the home. The tuner can be mounted in a bookcase, special cabinet, in a closet, or any convenient place, while the speaker can be concealed beneath a stairwell, behind false pictures, etc. A few typical installations can be seen in Altec Lansing's special demonstration room (see Fig. 2). In (a), the tuner is built into a chairside cabinet. The record changer has been mounted in the front portion of the cabinet and pulls forward. In (b), a duplex speaker is installed in a false corner built into one corner of the room. The speaker is located at an optimum level of about 5 feet above the floor. (The false corner provides a rear cavity in excess of 10 cubic feet for proper loading of the low-frequency cone.) It can be constructed in any room, and can be designed to hide vertical steam pipes, water pipes, wires, etc. (The tuner and record player can also be located in this false corner if desired.) In (c), the speaker is mounted in the wall behind the knick-knack shelf. The back of the shelf is a porous fabric (selected to match the wall) which totally conceals the speaker. Optimum tone quality will result if the back of the speaker can radiate into a large closet of several hundred cubic feet. The speaker in (d) is installed in a bass-reflex cabinet which acts as a functional piece of furniture. Another good installation is shown in (e). The entire system is mounted behind a closet door. This type of installation is particularly suited to apartments and small homes where space is at a premium. The speaker is mounted in the top half of the door and covered with porous cloth; the tuner, amplifier, and record changer are mounted behind the center panel. (This panel opens forward to give access to the controls without the necessity for removing the cabinet.)

In clothes closets, no sound-insulation material need be provided, since the sound absorbing qualities of the clothing will usually suffice. In installations where bare surfaces exist behind or to the side of the speaker, sound treatment must be applied. Fiberglas (available at refrigeration supply houses) or rock wool (found in hardware stores and lumber yards) are very good sound-deadening materials.

The sidelines mentioned in this article are only a few of the means which can be used to expand shop income. Their use has been the difference between success and failure of more than a few shops. The door to success can be opened—all you have to do is push!

For more than 6 years RADIO-ELECTRONIC ENGINEERING, a special edition of RADIO & TELEVI-SION NEWS, has kept alert engineers dependably and thoroughly informed on all that's really important in electronic engineering.

Selected and exclusive articles and other specialized features of practical and lasting value to electronic engineers are added to the content of the regular edition of RADIO & TELEVISION NEWS.

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New subscribers please use the lower half of order card in this issue.

The RADIO-ELECTRONIC ENGINEERING edition is available only by subscription.



D.C.-A.C. Inverter

(Continued from page 65)

amplifier having considerable voltage gain, it will be desirable to regulate the a.c. supply voltage. A simple way to do this is to connect a ballast unit in one of the supply leads.

For best results, the output load device operated from the inverter should be of high enough impedance to offer negligible loading. Operation of each of the inverter circuits given in this article is not noticeably affected by impedance of the d.c. signal source.

Outlook

We believe that the saturable reactor-type of d.c.-a.c. inverter and the magnetic amplifier merit consideration as useful components in amateur experimental circuits. Many interesting experiments and much practical development work can be carried on with these units. The units shown in this article have utilized readily available transformers having easily saturated cores. However, the powerhandling ability of these simple units is not great. Other experimenters may elect to build saturable reactors with high-permeability core materials and thus obtain power amplification in addition to voltage amplification. -30-

SYNC LOCK FREQUENCY CONTROL

By MATTHEW MANDL

TELEVISION technicians often find that they are called to service sets which only require readjustment of rear panel controls because the latter had been tampered with by the set owner. The control which has the most serious effect on the picture is the horizontal sync lock frequency control because a maladjustment of this will result in loss of sync to such an extent that the horizontal hold control is no longer able to bring the picture back into lock. On modern receivers this control is recessed or located in such a position that it is not easily accessible to indiscriminate handling by the set owner. On many of the older sets, however, the frequency control is located on the rear panel of the chassis and can be turned without the necessity for special aligning tools. Typical receivers are the Admiral 30A-I chassis, the RCA 630TS receiver, and the many kits and modifications of the popular 630-TS.

When servicing such receivers the technician can reduce the danger of tampering by covering this control with several strips of Scotch tape after he had made the proper adjustment. Inasmuch as the inherent stability of the sync lock system is high, this control will not need readjustment again unless a tube or component part of the horizontal sync system is changed. Some technicians use a pair of diagonals to cut off the knurled end of this control so that subsequent adjustment would require the use of a pair of pliers. This procedure will also tend to discourage tampering and help in cutting down on the number of unnecessary calls on the technician's list. -30-



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- TELEVISION SERVICING
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Stancor transformers are original components in thousands of radio and television receivers made by the biggest names in the industrythey have to be good to be specified by critical design engineers and value-conscious purchasing men!

Why shop around? Specify Stancor for your replacement work. You will get a dependable, honestly-rated transformer. You'll cut down on expensive call-backs. You will keep your customers.







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155



Again Atlas leads the field with its new line of speakers and projectors. Again Atlas makes the news in the Sound System field.



Atlas Reproducer units continue to retain the famous "Atlas V Plus" super-efficient magnetic assembly and in addition many more "Extra Plus" features. A new reversed dome, blast proof diaphragm is now standard in the high power, high fidelity models. Built-in transformers, designed for either constant 70 volt or constant impedance audio circuits are included. Improved appearance—functionally designed for maximum convenience. Double seal weatherproofing. All this and more without any general increase in price.

Atlas projectors have a new micrometically calculated and controlled rate of expansion. Atlas non-vibrant prajectors are rugged and fine in appearance. Sound energy is not dissipated in rattle vibration, distortion or cancellation.

The new improved line of Atlas speakers are really new from the voice coil to the final lock washer. It's really the "modern look" in speakers, a new high in overall performance.

Let Atlas speakers play an important part in your SOUND PROFITS.



Manufacturers' Literature

Beaders are asked to write directly to the manufacturer for the literature. By mentioning RADIO & TELEVISION NEWS, the issue and page, and enclosing the proper amount, when indicated, delay will be prevented.

WOODEN BEARINGS

A new engineering manual which covers the principles and applications of "oilless" wooden bearings in modern machine design has been released by the *Paramount Oilless Bearing Co.* of Worcester, Massachusetts.

The booklet discusses the application of wood bearings in certain electrical and electronic situations where no other type of bearing can serve as well. Typical applications include their use where the dielectric qualities of the bearing itself are important, where oil drip or spatter from conventional surface-lubricated bearings would injure the efficiency of surrounding circuit components, use in locations of difficult or impossible access, where surrounding high voltages prohibit lubrication maintenance, and where elimination of noise is paramount.

Copies of the manual will be loaned to accredited engineers and machine designers. Application for a copy of the manual should be made, on business letterhead, direct to the company.

ELECTRICAL CONDUITS

An elaborate, 30-page catalogue, the No. 603, covering the various types of electrical conduits manufactured by the company has recently been issued by *National Electric Products Corp.*, Chamber of Commerce Bldg., Pittsburgh 19, Pa.

The first half of the new booklet is devoted to Sherarduct, Xduct, and Economy rigid steel conduits and their accessories. The electrical metallic tubing produced by the company is next listed along with related accessories. A line of flexible steel conduit for branch circuit wiring and the company's non-metallic tubings are also described and illustrated.

AMERTRAN BULLETIN

A bulletin describing a redesigned and improved line of air-cooled transformers has been announced by *American Transformer Co.,* of 285 Emmet Street, Newark 5, New Jersey.

Designated bulletin No. 110-02, the publication contains 8 pages of information and data, including tables which list standard ratings with dimensions, weights, list prices, and wiring diagrams of transformers in capacities up to 200 kva., single phase, and 300 kva., three phase.

TRANSFORMER DATA

An 8-page folder containing 36 photographs and corresponding specifications on auto and insulated transformers has been released by Nothelfer Winding Laboratories of Albemarle Avenue, Trenton, N. J.

Illustrated and described are open frame, cased, channel, and end bell cased electrical transformers from .01 to 300 kva., 25 to 400 cycles, in 1, 2, or 3 phases.

In writing for copies, please specify *NWL* Bulletin No. 150.

POWER SUPPLIES

A new 4-page booklet describing its line of electronically-regulated power supplies has been issued by *Furst Electronics* of 12 S. Jefferson Street, Chicago 6, Illinois.

Various units are pictured and described in tabular form. In addition, details on the Model 910 klystron power supply are given. Copies of this pamphlet are available on request.

PARTS CATALOGUE

Radio Shack Corporation of 167 Washington Street, Boston 8, Massachusetts, has announced its new 152page parts catalogue for the trade.

Covering the fields of industrial electronics, communications, television, high-fidelity music systems, recording, testing, and research, the new catalogue lists complete equipment, kits, components, hardware, tools, test equipment, and technical books.

Copies of the catalogue are free of charge on request.

TV SERVICE DATA

A series of technical bulletins designed to assist the television technician has been issued by *American Phenolic Corporation* of 1830 South 54th Avenue, Chicago 50, Illinois.

One set of data sheets discusses the television antenna in some detail and provides valuable information on installation and orientation. The second booklet, entitled "Sell a TV Picture" is designed as a guidebook for personnel selling television receivers.

Copies of these helpful publications may be secured from the company on request.

MEISSNER KITS

A 4-page illustrated folder covering five new radio kits recently introduced by the company is now available from *Meissner Mfg. Division, Maguire Industries, Inc.,* Mt. Carmel, Illinois.

The booklet describes the T8CK FM receptor, the T4AK 20-watt amplifier kit, the T2BK 2-tube battery trainer kit, the T3BK a.c.-d.c. trainer kit, and the T6BK 6-tube a.c. superheterodyne kit for instructional purposes.

A copy of the new publication may be secured by writing the company and requesting the "New Radio Kits Folder."

WIRED TELEVISION

A booklet describing the operation and application possibilities of the "Utiliscope" has been issued by *Diamond Power Specialty Corporation* of Lancaster, Ohio.

The photographs and text material point out the varied ways in which wired television may serve industry and trade. The equipment is described as being particularly suitable for locations where the operation to be observed is inaccessible due to dangers involved, poor location, or adverse atmospheric conditions.

A brief description of the equipment required for the closed circuit television setup is given in the folder, copies of which are available on request.

AUDIO EQUIPMENT

The Electronic Workshop, Inc. of 351 Bleeker Street, New York 14, is currently offering two data sheets to persons interested in audio equipment.

One sheet covers the company's amplifier Models S-20 and A-20-5 while the second sheet describes the Series 510-A miniature audio oscillator. Performance features and technical data are provided on both of these units.

"DATALOG"

Electrical Reactance Corporation of Franklinville, New York has recently issued a "Datalog" covering its line of "HI-Q" capacitors, resistors, and choke coils.

Designed for the television and radio manufacturer and the producer of electronic devices, the new publication provides technical data in addition to the usual product information found in catalogues.

The "Datalog" carries 32 pages and is available on request from the company.

EQUIPMENT CATALOGUE

A catalogue covering the company's line of equipment has been issued by *Approved Electronic Instrument Corp.* of 142 Liberty Street, New York 6, New York.

Included are the company's 15" custom built TV receiver chassis, television field strength meter, sweep signal generator for TV and FM, marker generator, a signal generator kit, a sweep generator kit, FM tuner kit and tuner, a TV tuner, a 4 watt amplifier, and a 4 tube miniature broadcast superhet tuner.

"HAM NEWS"

As a service to radio amateurs throughout the country, the *General Electric Company* has announced that "Ham News," a bi-monthly publication currently being distributed to 65,000 amateurs and experimenters through the company's tube distributors, will now be available on subscription to those who find it difficult to obtain copies each month.





• Let it rain or blow—freeze or thaw—from arctic cold to blistering sandstorms, CLETRON loudspeakers keep functioning properly on P. A. systems and outdoor theatres with full-tone reproduction. CLETRON speakers have passed a 200-hour salt spray test and then another 200 hours on a weatherometer. In fact, these speakers actually operate when fully submerged in water.

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the CLETRON speaker to be the only one on the market to meet our severe requirements. Actual field experience from Cape Hatteras to the sands of Texas has proved that our faith in your product is entirely justified." E. B. Brady, Chief Engineer, Drive-In Theatre Equipment Company, Inc., Cleveland.

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PYRAMID CAPACITORS

Pyramid Electric Company of 155 Oxford Street, Paterson, New Jersey, has announced the availability of copies of its new 1950 catalogue which has been designated J-6.

The capacitors listed in the catalogue have been chosen as representative of the company's line and those which will adequately meet most of the technician's replacement requirements.

Units for both radio and television applications have been included.

"HISTORY REPEATS"

The Espey Mfg. Co., Inc. of 528 East 72nd Street, New York 21, New York, is now offering a new brochure, en-titled "History Repeats," to counter salesmen and sound salesmen of radio parts distributors and dealers.

Designed to familiarize sales personnel with the types of high quality sound equipment available for custom installations and the latest AM-FM radio receiver combinations, this publication lists a tuner, recorder, record changer, speaker, television chassis, and cabinet. The products of six companies are represented and are available as a "package" for custom jobs.

ANTENNA BOOKLET

An informative 16-page booklet on television and FM antennas, entitled "The Story of the Magic Wand," has just been released by Ward Products Corporation of 1523 East 45th Street, Cleveland, Ohio.

The new publication, which tells about the development and application of antennas, is available to television technicians and the general public.

Copies of the booklet are available without charge from the company.

INSTRUMENTS MANUAL

A new and revised edition of the "Manual of Electric Instruments" has been issued by the General Electric Company's Meter and Instrument Divisions.

The manual describes the fundamentals of construction and operating principles of all major types of electric instruments. Written primarily for use in schools, the manual may be used by men in the industry who wish togain familiarity with the electric instruments with which they work. The thirteen chapters of the book explain the fundamentals of such instruments as thermocouples, synchroscopes, frequency meters, and electric telemeters.

The book contains 150 pages and the text material is lavishly illustrated with photographs and diagrams. Copies of the publication, GET-1087A, are available from General Electric Company, Schenectady 5, New York. The price is \$1.00 a copy. -30-

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International Short-Wave

(Continued from page 130)

French Equatorial Africa—Pearce, England, says Radio Brazzaville, 11.972, 9.440, now has news for Europe at 1315-1330; at close announces English daily at 0115, 0510 in the 19-m. band, and 1315 and 1550 in the 25and 31-m. bands.

French Indo-China—Balbi, Calif., reports the "Voice of Vietnam" now heard on 9.670 and (new) 7.265 from 0500; news 0615.

Although Saigon was to have moved some time ago from 11.78 to 9.524, the station has not been reported to me recently as heard on either of these channels; should have news 1845, 1945, 0500, 0900.

French Morocco-Rabat, 6.006, noted 0200-0230 in French; man announcer identified as "Ici Radio Maroc" at 0230; music followed. (Cox, Delaware) Peddle, Newfoundland, confirms overseas reports that Rabat also is being heard on approximately 8.190 afternoons (EST); spurious?

French West Africa—According to its verification card, *Radio Dakar* now operates on 11.895 at 0200-0330, 0700-0830, 1320-1800, and on 15.340 at 1400-1530 daily. (Radio Sweden)

Galapagos Islands—The Swedish Fan DX Club reports a *new* station giving location as Santa Maria, Ecuador; language is Spanish but often has English announcement; operates on approximately 4.960; it is believed this may be a *new* outlet on Santa Maria Island, located approximately 800 miles west of Ecuador in the Galapagos Group. (HC5HP, Riobamba, Ecuador. is listed on 4.960 with 250 watts.—KRB)

Germany—The two short-wave transmitters of Der Deutschlandsender, according to announcement, are operating on 6.115 and 7.150; schedule appears to be 0400-1800 daily. (Radio Sweden) The 6.115 outlet is heard at good strength in Britain; news in German 1500; appears to be Soviet-controlled, and is believed to be "Rundfunk, Berlin." (Pearce)

Hamburg, 7.29, sends a strong signal to the British Isles; noted with news in German 0645 (except Sun.), also at 0245, 1345, 1800; sends usual letter in German and view of transmitters, lists power 50 kw. (Pearce)

Greece—"Free Greek Radio" has been widely reported daily 0045-0100 on 9.460; language is Greek; since this is a clandestine outlet, definite location is not known but it is believed to be behind the Iron Curtain. The station is now in its third year of operation and formerly was in the 6- and 7-mc. bands. A late report from Stark, Texas, indicates the station may have moved or left the air, since he had not heard it more recently at 0045-0100.

Bluman, Israel, has checked a num-

ber of Greek regional s.w. outlets with these results-6.345, a station relaying Athens II on 1550 kc., at 0000-0200, 0500-0800, 1100-1545; 6 530, Janina, at 0000-0030, 0700-0730, 1200-1500; 6.745, Larissa, at 0000-0200, 0430-0800, 1000-1500; 7.020, Radio Peloponesos, at Tripolis, at 0000-0200, 1100-1430, barely audible 0700; 7.050, Athens, Forces Station, at 0000-0230, 0430-0730, 1000-1500; 9.607, a station relaying Athens I on 601 kc., at 0030-0230, 0500-0630; these are weekday schedules, additional transmissions are carried by all stations on Sunday. (Radio Sweden)

Guatemala—At the time this was compiled, TGWA, 15.17, and 9.763, was still announcing as "La Voz de Guatemala." Press dispatches early in the year indicated that the country's name was being changed to Isthmania. Is still verifying with an "ancient" QSL card, almost 20 years old.

TGWB, Guatemala City, noted moved back to measured 6.4396 from 6.07; signal reported good evenings except for some QRN. (Treibel, Washington)

Haiti—At the time this was written, 4VRW, Port-au-Prince, was using approximately 10.207, but still announced 10.130; it is possible that the English announcement is a recording. (Hankins, Pa.) On Friday nights at 2000-2030, 4V2S, 5.951, Port-au-Prince, has an *English* session with tourist information. (Ferguson, N.C.)

Hawaii - KRHK, 15.25, Honolulu,

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has replaced KRHO, 9.65, to Far East 0400-0900. (Balbi, Calif.)

Hong Kong—ZBW3, 9.525, is heard to closing 1030 with playing of "God Save the King." (Radio Australia)

Indonesia—Direct from Djakarta, it has been confirmed that the station heard 1115-1400 on 11.785 is YDF2; this is the *new* 100 kw. transmitter which should be testing soon on 9.520, probably at 1000-1100. (Moore, Calif.) By this time the 11.785 channel may have added *English* for Europe at 1400-1500; and the English period (on YDF, 6.045, when this was compiled) may have moved by now to YDF2, 11.785, or possibly to 9.520 (on which tests were to be effected soon).

Fuller, R. I., is hearing the old outlet on 19.345 (formerly Batavia, now Djakarta) at 1130-1200.

A station heard recently by Stark, Texas, on approximately 9.665 at 0800 in Dutch may be Menado, Celebes. *Radio Australia* reports this station, YDS, on a *new* channel of 9.680, heard with weak signal as early as 0530, but at sufficient strength by 0600 to read announcement, "Hier ist Menado, Radio Indonesia"; orchestral items 0530-0600, then news in Dutch.

NNRC reports YDQ3, 11.084, Makassar, Celebes, heard 0745 with recorded dance music and with *English* announcement 0800, followed by news in Dutch.

Iran — *Radio Teheran*, 9.66, noted 1345 with *English* announcement, playing classical selections; news in French 1400-1407, then back into *English*, playing popular music. (Boice, Conn.)

Iraq—Baghdad II, 7.063, signs on 0830 with program in Kurdish, which sounds like Persian; at 1115-1200 has *English* program. (Bluman, Israel, via Radio Sweden)

Ireland—At the time this was written, Radio Eireann's new 100 kw. transmitter had not yet been reported as testing. When the opening date is known, schedules will be released through Irish Legations; recently listed s.w. schedules for the low-powered (1.5 kw.) transmitter at Athlone were for news at 1330-1350 on 17.840, and at 1710-1730 on 11.760 (officially listed 11.74); the 25-m. band outlet is not heard in America as far as I know, but occasionally I have heard the 17.840 outlet through terriffic QRM from Radio Brazzaville.

Israel—Tel Aviv's listed 9.000 outlet recently has been measured 9.010. (Ferguson, N. C., others)

Japan—AFRS, 9.605, Tokyo, noted in Delaware 0300-0330. (Cox) WLKS, measured 6.115, Kure, noted 0600 with BBC news relay. (Triebel, Washington)

Lebanon — Radio Liban (formerly Radio Levant), Beirut, is scheduled daily 0000-0130, 0600-0730, and 1000-1535 on 8.036. The station says the name change was effected when the station was handed over to Lebanese authorities on April 1, 1946. (Radio, Sweden)

Madagascar-Radio Tananarive,



RADIO & TELEVISION NEWS

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9.695, noted 2300 with news in French. (Stark, Texas) Heard in Ohio as early as 2230. (Sutton)

Malaya—Radio Malaya, 9.712, noted to closing 1030, good signal in Calif. (Rosenauer) The Blue Network is now heard in dual on 7.200 and 7.250; news 0900. (Dilg, Balbi, Rosenauer, Calif.) While broadcasts are in Malay and Chinese dialects, the stations can be identified when they announce "Radio Malaya." (Radio Australia)

Malta — The Forces Broadcasting Service, Middle East, is being heard well in East from 2330 sign-on on both 4.965, 7.270. Verification card lists frequencies allocated as 7,220, 7.270, 4.965, 11.785, 6.140, all 7.5 kw.; former 4.782 channel is crossed out; hours of broadcast are listed 2330-0200, 0430-1700.

Mauritius - V3USE, approximately 15.050, is being heard erratically in the Eastern U. S. at 2159-2315 (approximately), and in both East and West USA around 0930-1200; has BBC (*English*) news relay 2200, and sometimes carries "*English by Radio*" shortly after the end of the news; most other items are in French; frequency seems to vary.

Mozambique-At the time this was compiled, Lourenco Marques seemed to be experimenting in its English language program-was being heard weekdays from 2300 and Sunday from 0000 on 11.764, 9.763, 6.137 (all approximate). August Balbi, Calif., reports this is the first time Lourenco Marques has been noted in the 25-m. band since 1937; widely reported; announces that the 25-m. outlet is for "listeners in the Cape." This outlet has been measured 11.7692 by Treibel, Washington. More recently the 31-m. outlet has been on approximately 9.755, probably having dropped to avoid QRM from TGWA and Leopoldville.

The Portuguese program was noted recently in Pennsylvania coming on the air 2300 on approximately 9.705, 9.660, 9.568; 9.660 was best level. (Hankins, Pa.)

Peddle, Newfoundland, says the best signal from Lourenco Marques is now over CR7BG, 15.196, heard first testing on Thursdays to 1500; later has been heard on other days; some days may run to 1530. Announces as beamed "regularly" to Europe and Portugal and Portuguese Colonies; always gives program preview for the following day. I have heard this one at good level here in West Virginia as early as 1400. Noted in England by Pearce on occasion after 1505 with messages addressed to Lisbon, Angola, and other Portuguese possessions; signed off 1525 with the Portuguese National Anthem, "A Portugesa"; Pearce says this appears to be the same program as radiated on 4.825 and also sometimes on approximately 9.670.

New Caledonia — Slutter, Pa., is hearing a station on 6.000 signing on 0230 with "La Marseillaise"; probably is Noumea; has music to 0245.



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New Zealand-ZL3, 11.78, ZL4, 15.28, now parallel at 2300-0635 sign-off; ZL7, 6.080, is now off the air. (Balbi, Calif.)

Nigeria-A station heard by Dilg, Calif., on approximately 9.655, with irregular program format, may be Radio Nigeria, Lagos. Noted with native singing both before and after 1100.

Radio Nigeria verified by airmail from the Director of Post and Telegraphs, P&T Headquarters, Catholic Mission St., Lagos, Nigeria. By now should be operating 0100-0215, 0600-1700 on 7.255 with 1 kw.; 0100-0215, 0600-1300, on 9.655, 300 watts, 1300-1600, on 4.990, 300 watts; on Sunday should operate 0100-1700. The "morning" sessions are relayed from the BBC's General Overseas Service: native programs are radiated "evenings." Broadcasts are still experimental, quality is not the best. Transmitters are Air Ministry T1509, 300 watts, and RCA 4331, 1 kw. Nigeria plans new stations at Ibadan, Kaduna, and Enigu for 1951. (Cushen, N. Z., via URDXC)

Outer Mongolia-Ulan-Bator, 8.400, noted with oriental-type program to 0800, then with Moscow relay. (Rosenauer, Calif.) Noted from 0400, fair to excellent. (Balbi, Calif.) Pakistan—Radio Pakistan, Lahore,

has been radiating on 6.075 since November 1, 1949. (Sampat, India)

Frequencies and schedules of Radio Pakistan seem to be irregular, but at the time this was compiled English news periods appeared to be radiated on 9.646 at 2145 and 1015; on 11.885 at 0210; on 11.845 at 0700 and 0820; on 17.825 at 0110 and 0210; on 11.548 at 0700, 0820, 1015; Dacca was using 7.635 for English news at 0700 and 1015, and may use 15.335 for English news at either 2100 or 2145; however, there is a possibility that the 15.335 outlet heard evenings (from around 2045) may be Karachi. There definitely are two 50 kw. transmitters now operating from Karachi. It is possible, also, that at times Dacca may be using its old channel of 15.27. (Hankins, Pa., Morris, N. Y., Sampat, India)

Panama-A station announcing "Radio Continental," located in Panama City, has been heard with news in Spanish ending 1930 on 6.0445; also noted at 0630. (Oskay, N. J.) Noted in Washington signing off 2400. (Treibel) Radio Sweden says this station's QRA is believed to be Radio Programas Continental, S. A., Apartado 1795, Panama City, Republic of Panama.

Peru - OAX5C, measured 9.4776, Ica, noted 2315 with classical music. (Treibel, Washington)

Philippines - Balbi, Calif., is still hearing DZH7, Manila, on 9.730, where it is QRM'd by BEA8, 9.735, Nanking; noted in English 0400-0500, news at start. Balbi notes that DZH3, 9.50, Manila, has news 0400, 0500.

According to the "Philippines Free Press," weekly news-magazine, the Bacolod Broadcasting Corporation is operating DYBR, 1120 kc., and DYB2, 4.985; this is Bacolod's first station and was formally inaugurated December 25, 1949, although had tested



than in any other magazine. Monthly net paid circulation over 200,000.

from October 13, 1949. Geographically, Bacolod is 300 miles SSE of Manila. (Loboyen, Ill.) Cushen, N. Z., says DYB2, "The Voice of Bacolod," 4.985, operates with 250 w., and is heard to 0900, but has bad CWQRM; that Davao is on 3.950 with no call vet assigned, and that another station DXAW, "The Pioneer of Davao," operates on m.w. 1180 kc., only, as yet; that DYRC, 600 kc. and DYH2, 6.140, "The Voice of Cebu," 151 Colon St., Cebu City, operates 1600-1100, station being owned by the Cebu Broadcasting Co., operated by the Manila Broadcasting Co., and affiliated with NBC; DYH2 is heard on Sundays at 0530 with a "Voice of America" transcription, "Make Believe Ballroom," Cushen says.

Pitcairn Island — Overseas sources report vaguely that a station, XKG, is operating from this island on 8.290; no other details.

Portuguese Guinea - CQM7, 6.993, Bissau, excellent in Newfoundland 1630-1800 closedown. (Peddle)

Portuguese India-Radio Goa, 9.610, now broadcasts in English daily; the schedule over-all is 2030-0640; 0730-0930; program consists of light and classical music on records and request program. It also uses Portuguese, Konkani, Hindi, Urdu, Sindhi, Marathi, and Gujarati. On Wednesdays carries English religious program ("Bringing Christ to the Nations").

Roumania-Radio Bucharest, 9.252, usually is readable, fair level, 1500-1530 in English.

In verifying for Ferguson, N. C. Radio Bucharest listed channels of 823 kc., 5.990, 6.205, 9.250, 11.900.

Saudi Arabia-A Danish DX-er reports Mecca on 11.950 at 1200-1317: a German listener has measured frequencies as 5.976, 9.652, and 11.750, heard in Arabic 1200-1300. (Radio Sweden) Cushen, N. Z., has heard Mecca testing 1300-1315 in English on 3.950. and then on 3.060 when changed from rhombic to doublet antenna; also heard on 5.990, 9.650, and 11.750, according to announcements, these are tests for reception in Cairo; tests 1200-1300 in Arabic. (Radio Australia)

South Africa - The South African Broadcasting Corporation is to have a new transmitter in service from May 1. (Radio Sweden) Other sources indicate that perhaps more than one new transmitter will be put in operation. The new installation is for the "C" (commercial) program to be called "Springbok Radio." The service will operate on s.w. in the 31-, 40-, 60-, 90-, and 120-meter bands daily at 2345-1600 and Saturdays to 1200; exact frequencies to be used were not known at the time this was compiled.

According to Short Wave News, London, ZRB, Air Force Station at Pretoria, now operates on only 9.110 on Sunday-Thursday at 2345-1015 and Friday 2345-0500; verifies with card; QRA is Officer Commanding No. 64 Air School, P. O. Odonata, District of Pretoria, Transvaal, South Africa.

ZRB, measured 6.1096, noted 0020

May, 1950



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RADIO & TELEVISION NEWS

4

been unable to hear any such station. If anyone has actually "heard" this outlet, please let me know.

According to Lowell Thomas, the American newscaster, "Radio Tibet" is being operated by Reggie Fox, AC4YN in English, Chinese, Tibetan. Fox, formerly a representative of the British Political Mission, resigned his post, married a Tibețan girl, now holds a high post in the regime of the Dalai Lama, and is now considered a native of Tibet, it is reported. According to Mr. Thomas, Fox expects to spend the remainder of his life in Tibet. Two of his four children are now attending school in India; they are being trained to fill high government posts in Tibet. (DeMyer, Mich.)

USSR - A Soviet outlet noted in Russian on 7.93 from 0600, powerful signal; the 9.48 outlet heard with Home Service from 0200, strong signal in California. (Balbi)

Vatican-HVJ, 7.28, noted in Sweden with Spanish 1500. (Aberg)

Last Minute Tips

A station reported earlier by *Radio Sweden* as "unknown" has now been identified as *Radio Euscari*, "La Voz de la resistencia Basca"; this clandestine outlet operates on 6.090 at 0200-0230, and on 6.525 at 1600-1630. (Radio Sweden)

A transmitter heard by Bluman, Israel, on 7.075, is probably Bahrein Island in the Persian Gulf; heard before 1050 when the station is blotted out by Ramallah's carrier. (Radio Sweden)

At the time this was compiled, Boice, Conn., reported Paris, 9.55, 9.68, had discontinued the *English* beam to North America 1945-2000.

Radio Sweden says the Lisbon outlet formerly on 15.160 has now moved to 15.147; also reports Portuguese Radio Clube, Oporto, on 7.530 at 1300-1800.

A station heard by Peddle, Newfoundland, on 11.950 at 1230-1300 is believed to be Mecca, Saudi Arabia; poor level. Pearce, England, is hearing this one at 1200-1300, but occasionally continues to test to around 1315 when announces "This is Saudi Arabia Broadcasting System."

A strong station heard on 7.906 at 1200-1300 is presumed to be a *new* Greek outlet. (Radio Sweden)

The Forces Station, Central Greece, Larissa, noted back on approximately 6.745 from 6.770; sent QSL card; listed *English* program for Thursday 1530. (Pearce, England) Gives schedule as 0000-0200, 0500-0800, 1100-1545. (Radio Australia)

PRL8, 11.72, Rio de Janeiro, Brazil, now heard with fine level to 0355 signoff; this transmission is relay of m.w. "Radio Maua." (Worris, N. Y.)

I recently noted *Radio Brazzaville*, 11.972, on a Sunday at 1100 with news for the Far East; fair level; announced was operating in the 25-, 31-, and 48-m. bands.

According to Radio Sweden, FBS, Middle East, Malta, has been heard

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on a new channel of 9.925 around 1500. Fargo, Ga., recently heard a station on approximately 8.87 with news 1530 announcing as Jerusalem, Israel.

*

* **Press Time Flashes**

*

Direct from Leslie Knight, Chief, Forces Broadcasting Service, Middle East, Malta, I learn that later in the year FBS may inaugurate a special weekly program for SWL's; if so, most likely will be around 1715 EST (2215 GMT). For the present, schedules are 4.965, 2300-0200, 0830-1700; 7.270, 2300-0200, 0900-1700; 11.895, 0430-0830.

YDC, 15.15, Djakarta, noted good level 0900-1000 in English to China and Burma; announces YDB2, 4.91, in dual; same channels are used for 0600-0700 English period; during each English session has news at start and just before close. Radio Sweden confirms that European beam from YDF2, 11.785, has been extended to 1115-1500 and that last hour is English.

Verifying for DeMyer, Michigan, Mauritius, listed frequency of 15.075 (actually heard on approximately 15.05) and schedule with 1.5 kw. as 2200-2315, 0300-0430, 0930-1200; is best here 2300, usually leaves air around 2310 to 2320 with playing of first half of "God Save the King"; stated antenna is delta-type; uses both French, English.

Boice, Conn., reports Radio Pakistan, 11.885, Karachi, with news at dictation speed for Middle East at 1330; confirmed here, good level; seems daily feature now.

QRA for "Radio Elizabeth," OQ2AB, 11.900, 7.150, is A.F.M. Schovens, Box 1039, Elizabethville, Belgian Congo; QRA for OQ2AC, 3.390, 4.980, 7.200, is College des Francaise, De Sales, Elizabethville, Belgian Congo; both send excellent QSL cards. (Nordh, Sweden, via ISWC, London)

BCAF, 8.996, Teipeh, Taiwan, noted in Australia 0630 with English-Chinese lesson; Western music 0715. Bangkok, 6.24, 1591, noted both 0515, 0615, with news. DYB2, 4.985, Bacolod City, Philippines, heard 0530 with world and local news, then music. ZJM7, 11.72, Cyprus, noted 1600 with Arabic news, then music. (Sanderson)

Short Wave News, London, has announced it shortly will publish a successor to its "Short Wave Listeners' Annual" under the title, "The Op-Aid"; will include prefix lists, both by country and by prefix, zone boundaries, time tables, postage and mileage data, and much other essential information in a form which will aid ready reference. But will not include s.w. broadcast station lists "as this section is so well covered in the World Radio Handbook for Listeners."

The Overseas Broadcasting Station, Bangkok, Thailand, verified for Bellington, N. Y., with attractive card.

Serrano, Brazil, flashes that Lourenco Marques, 15.196, has news in Portuguese 1320, 1450; last news is followed by program preview; schedule seems 1300-1500 although some days runs as late as 1530.



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DUH5, 11.84, Manila, heard weak around 0230 in Oregon. (Callarman)

Radio Brazzaville now has news 1745 on 11.972 to North America; on 9.440 and 9.987 to South America; first two best here in W. Va.; seems to have dropped newscasts formerly at 1900 and 0030.

A station reported widely on approximately 9.530 around 0000-0130 is believed Warsaw.

OXI, 5.940, Greenland, noted 1735 in Danish; off 1748. (Bromley, Ontario)

A station that may be DYH3, Manila, noted by Dilg and Balbi, Calif., on approximately 6.11; announces as "largest station in the Philippines"; good 0500.

Saigon, French Indo-China, noted by Balbi, Calif., on 11.84, a *new* channel, from 0355; confirmed here in W. Va. at 0500 with news by woman; at 0515 was buried by LRT, Argentina, same frequency. Noted by Balbi on "old" 11.78 channel at 1015 with French news relay from Paris, signed off 1030. I cannot find it anywhere at 1800 former opening time on 11.78.

The Radio Club of Sweden says Damascus, 6.000, Syria, now has news 1530.

CR6RO, 7.582, Radio Clube do Bie, Silva Porto, Angola, says power soon will be increased from 400 w. to 1 kw. *Radio Omdurman*, Anglo-Egyptian Sudan, is on 6.115 now in addition to 9.747; has *English* on Friday 1230-1300. (Radio Club of Sweden)

YDQ3, 11.084, Makassar, Celebes, noted 0745; announced in *English* 0800. (Sutton, Ohio)

Fargo, Ga., flashes he has picked up Israel on 9.000 with news 1700, then a question-and-answer program about Israel; gave QRA as P.O. Box 754, Jerusalem, Israel, Station IBX; excellent level, in clear.

Moscow announced current daily schedules to North America are 1820-1930, 15.23, 11.88, 9.72, 9.67, 7.36, 7.29; 2000-2100, 15.23, 11.88, 9.72, 9.67, 9.60, 7.36, 7.29; 2200-2255, 15.41, 15.23, 15.10, 11.78, 9.72, 9.67, 7.29; however, 11.96 is also used in *each* of these transmissions. (Lane, South Dakota) The 11.96 channel is best here in West Virginia most evenings.

Radio Clube de Huambo, 11.925, Nova Lisboa, Angola, is scheduled weekdays 1300-1700, Sunday 0600-0745, 1100-1200. (Radio Sweden)

Radio Sweden reports Mafeking, Bechuanaland, has moved from 5.900 to approximately 8.230; on air daily 1200-1430.

Radio Tananarive, 6.160, Madagascar, heard in Calif. 0815-0915; music with announcements in French; call "Ici Radio Tananarive" is given irregularly; good level. (McPhadden)

Both Ferguson, N. C., and I recently heard a Chinese station on approximately 15.050 from tuning around 0610 to sign-off 0930; had Peking news relay 0830; good level; may have been testing.



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Aligning TV (Continued from page 44)

with the proper channel setting the raster should be observed on the screen. At this point the focus must be properly adjusted. Focus will vary with the position of the scanning spot on the screen although electromagnetic deflection exhibits less tendency to this than electrostatic. The trouble is nonetheless prevalent and must be constantly anticipated. The best picture focus is obtained when the scanning line is of the sharpest definition and the minimum width. This can only be observed upon close scrutiny of the face of the cathode-ray tube. Since focus also varies with grid voltage it is sometimes necessary to adjust the tube focus under conditions of extremely strong video signals which tend to defocus the beams.

At this point it is necessary to describe the correct determination of the proper number of lines in the vertical wedge. A proper wedge pattern terminates where the sharp definition between the lines and spaces of the wedge become soft or greyed in character. Slight readjustment of the focus control often proves that the trouble lies in the focus of the beam in that area. This is very important in determining the correct i.f. alignment.

Linearity

Vertical and horizontal linearity may be adjusted by observing the circularity of the test pattern's outer rings. Remember that both complete circles must be round. Adjustment of the linearity controls will compensate for any deviations from normal. If it is not possible to achieve a perfect circle then it is best to leave the setting at that point where the best figure is obtained. Nonlinearity will cause the test pattern to appear egg shaped. Failure to adjust for proper linearity will cause distortion of the wedges and less accurate setting of the video i.f. may result. The picture size should be adjusted for the correct height and width before checking linearity.

The brightness affects the picture focus and contrast. To adjust this control properly turn the brightness up until retrace lines appear, then decrease the brightness just to the point where they disappear. Too much brightness spoils the contrast effect and defocuses the beam.

Proper Video Bandpass

By carefully observing the test pattern the set can now be aligned. First observe the pattern. If, for example, the horizontal wedges are clearly visible only to the mark of 100 lines on the vertical wedge, it will be necessary to adjust the last i.f. to lower the i.f. frequency. Next the focus should be adjusted until the greatest number of vertical wedges can be



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12 TUBE

seen. If no change is perceptible it is advisable to let the i.f. remain in its original position and go on to the next stage. It may be noticed that although good vertical wedge information is present, the horizontal wedge information is smeared and very poor. This could be due to improper load resistors in the plates of the video amplifiers or improper setting of the i.f.'s controlling the low video frequency. Low video frequency means high frequency i.f.'s, therefore, the iron core slugs should be pulled farther out of the coil. When the bandpass is properly adjusted the vertical and horizontal wedges will appear in the same blackness and in sharp definition. The horizontal wedges will extend into the very center of the pattern and down to the inner circle whereas the vertical wedge may cut off at something greater than 300 lines.

It is a common receiver fault to receive a signal which at first appears as a ghost followed by other lighter ghosts which cannot be removed by orientation of the antenna. This may be caused by overpeaking the video i.f.'s. Further separation of the low frequency i.f. slugs will minimize this condition and even eliminate it in most cases. This condition is undesirable and is indicative of misalignment. It usually appears after a black line and the black line is followed by a white line which in turn is followed by a grey line, etc. These lines are caused by the damped oscillations set up in the video amplifier with a strong signal and the two i.f.'s too close together in frequency, permitting oscillation.

Low frequency distortion is caused by improper values in video amplifiers and results in blurring of horizontal lines even under fine-focus conditions. This condition is generally accompanied by excessive phase shift. If changing the frequency of the high frequency i.f.'s (low frequency video signals) does not correct this condition the video amplifiers should be checked since the r.f. head is not passing the 6 mc. required for good television re-Alignment must be perception. formed in the r.f. section of the head by adjusting the r.f. slugs in the coils for low frequency video.

As the vertical wedges become more sharply defined, sound bars will appear on the raster. These are usually horizontal grey bars which travel vertically across the raster. Elimination of these bars is necessary in order that the best bandpass may be selected. Adjustment of the sound trap is necessary. The bars may be eliminated by tuning the sound trap until they disappear or by presetting the sound trap by means of a signal from a signal generator tuned to the audio i.f. frequency. The same vertical bars will appear on the raster if the signal generator is modulated with 400 cycles. The sound trap is in tune when these bars disappear from the raster. It is obvious upon careful study of

May, 1950





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the r.f. bandwidth picture that as a higher and higher video response is approached the more sound will be allowed to enter the video i.f. The sound trap tunes very sharply but careful adjustment will always eliminate the sound bars from the picture.

Audio

The television audio uses a bandpass of 50 kc.--25 kc. on either side of the audio carrier. It is advisable to align a television receiver using audio as a basis for determining the best alignment point. The reason for this is that the video is 5.25 mc, wide and is difficult to set accurately even with a trained eye. However, it can be set within a few kc. by means of the audio since the greatest possible error with the audio is only 25 kc. or an error of less than .01 per-cent!

First align the audio i.f.'s by turning the slugs an equal number of turns until the audio is present in the speaker and the picture is visible on the screen. Next while observing the r.f. variable condenser or fine tuning control note the point at which the best picture is received and also the point at which the sound is heard best. If the capacitance or inductance increases its value the sound is below the picture by too great an amount and the audio frequency must be increased, i.e., the slugs must be pulled out of the coils. If the capacitance or inductance decreases when bringing in the best sound, the audio i.f. frequency must be lowered. After each adjustment check again with the highest definition picture to determine the new audio position and verify whether the point of best picture performance has been reached. Do not expect good sound until the discriminator has been properly balanced for each new setting.

The discriminator is balanced if the noise is at a minimum and the sound is not distorted on loud signals. This is important. Since the FM frequency bandpass governs the sound intensity, if only a weak signal is judged, very little can be determined about good FM balance. Test the audio on both weak and strong signals. It should handle both with no distortion and no noise. Although it may seem as if there were a loss in gain remember that a properly balanced discriminator has little or no background noise and handles full signal intensity without distortion. This point should also be the point of greatest video definition. When the audio and video are prealigned with a standard signal generator very little audio-video juggling is necessary. Remember that in using a standard signal generator the discriminator must be detuned and the i.f.'s lined up for greatest audio signal, then the discriminator is adjusted for weak or no audio signal.

Bandpass

At this point the receiver may be checked for bandpass. If the best possible picture has been achieved it

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may be assumed that the greatest available video band is being passed. Trouble may be present in the video amplifiers or the r.f. assembly. A simple tuning tool may first be tried by placing it at the peaking coils of the video amplifiers. There should be no change for the worse when the tool is brought near these coils. Another simple tuning tool may be constructed by fastening a peaking coil of 200 µhy. to a tuning tool. This coil may be used to shunt the peaking coils in the video amplifiers. This should reduce the picture resolution. If it increases the picture resolution the video coils are not properly designed or have changed value. A peaking coil of less inductance than was originally used in the video should then replace the present one if this effect is noted.

Usually, however, the trouble lies in the r.f. assembly. The bandpass of the r.f. assembly may be less than 6 mc. This is evidenced by good sound, good vertical wedges, but poor horizontal wedges. This may be checked by observing the test pattern as the fine tuning is varied. If the test pattern increases in vertical or horizontal resolution as the audio disappears or becomes louder, then all the required band is not reaching the i.f. amplifiers. This band may be increased by adjusting the r.f. trimmer slugs or condensers. Increasing the inductance for increased low video response and decreasing the inductance for increased high video response is the procedure to be followed. Remember that a good test pattern will show very clearly and have sharply defined lines and spaces of equal intensity both horizontally and vertically with no ringing evident and no phase distortion present.

The well designed receiver will produce over 300 lines vertically and horizontally. When this is achieved the receiver is properly aligned. All of this can be accomplished satisfactorily without employing elaborate test equipment—just the intelligent use of the video test pattern on the cathode-ray screen.

The cathode-ray tube is the handiest possible test instrument since it is always available during servicing procedures. -30-



May, 1950



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TSM-75K Super-Modulation Transmitter

A revolutionary new high power, compact, light weight amateur transmitter for stationary and mobile usc. The "TALK-POWER OUTPUT" is approximately 3 times that of any other type of transmitter with the same power input. Maximum power input of the TSM-75K is approximately 75 watts.

The basic kit now available consists of all metalwork completely finished, and includes punched chassis, front panel, back panel, cabinet, and all brackets and special parts. Also included in the kit are the modulation transformer, meter shunts, speclal capacitators, and all other special parts. The major feature of the TSM-75K basic kit is the cable harness, which is completely pre-formed and laced, and all wires are stripped, tinned, and numbered for easy assembly. The wires are placed in such a way that when the harness is laid into the chassis all leads are close to the parts to which they are to be attached. Complete assembly, wiring, and adjustment instructions are furnished, and form ished are standard parts available from most jobbers. Tubes required are 1 - 3E29/829B, 2 - 6AK6, 2 - 6AQ5, and 1 - 6AU6.

The modulator is incorporated into the transmitter, and the only accessories required for operation on either phone or CW are a microphone and power supply. Coil data is supplied with the kit for 10 meter operation. Other data available on request. Coils are available at prices shown below.

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MORE FORGOTTEN MEN

CDLEASE add my name to the list of those in 100 per-cent agreement with the sentiments voiced so ably by Mr. Van Slyck (February issue). Last November I became a charter member of the 'Forgotten Men' fraternity.

"I have handled the receivers of a prominent manufacturer for over three years. I have built them up to a position at the top of the customer acceptance list in this territory. Suddenly---no more radio sets. I lost about \$450.00 at Christmas and will lose much more in the months to come. I have written to nearly every vicepresident they've got and the best I have had is the wish that I will be able to hold out until 'television gets to Mena.' Mena, Arkansas, has a population of 5000 and is 85 and 95 miles respectively from the nearest FM stations!

"In the last three years I have sold nearly \$25,000 (list) worth of this manufacturer's products. They won't sell that much TV in the whole dadburned state of Arkansas in the next five years!

'I don't know what is going to happen to the organization of once-loyal dealers which has been built up so laboriously over the years. Certainly the companies are going to have to start all over again to get any goodwill out of us, if the rest are like me. My contention is this. Radio has a large listening audience. What percentage of these use a \$19.95 midget or \$39.50 table model? How many years will it be before TV advances to the point where they can provide the mass market with a table model selling for less than \$50.00? Einstein's new theory may provide the answer-I certainly can't."

Jack Darr

Ouachita Radio Service Mena, Ark.

THE GAIN SET

CEN READ with a good deal of interest the article on the audio oscillator and gain set in the January issue but would like to make a point.

"Mr. Bumbaugh leaves the impression that no matter what his gain set is terminated into, beyond the loading resistor, that he need not bother to include the loss of the termination in the total loss of the input attenuator. In other words, no mention is made when referring to Fig. 6, that when building out to high impedance, one must add the loss of the pad thus created to that of the gain set. He is not wrong, true, but too many people don't know the difference.

"In the case cited, adding the 100,000 ohms to an assumed 100,000 ohm grid (itself assumed as a pure resistance which is not so since there is always an appreciable capacity at audio frequencies in the measure band) would create an impedance ratio of 200 to 1 or an additional bridging loss of 26 db. This additional loss must be added to that of the gain set before we can evaluate the single frequency gain of the equipment under test. The same thing applies to the matching network for the low impedance case cited. I will say, however, that the low impedance case is a lot easier to evaluate than the high.

"Conventional practice uses two methods. One, a matching transformer of good design and two, the matching pad, a variation of which Mr. Bumbaugh has shown. I would like to see a clarification of the added loss made public just in order to keep a lot of well-meaning radiomen from running into trouble over something which is far from well-known even in the broadcasting industry.

"As a sidelight, I might add that I seriously doubt if there are more than two per-cent of the engineers presently using audio measurements in the broadcast industry who actually know how to use the decibel properly. Call any Master Control you like, even here in Chicago, and two-to-one when you ask for a level you will be told 'It's leaving here at 0 db.' Now you know that doesn't mean a darn thing. What is the reference point for such **a** statement? Take it further and ask the same engineer what reference he is using and I'll bet again he doesn't know. That has been my experience for years and I have fought it like mad. The average radio engineer has an idea that the db. is an actual value of voltage. That is the big reason for going to the v.u. system where the reference is part of the definition of the v.u. and where a reading of so many v.u. means what it says.

Hilton Remley Station WGN Chicago, Illinois

TV IN HIGH SCHOOLS

WE FEEL that you may be interested in knowing that television is definitely spreading into the high schools.

"Our senior class play was televised and aired on our local outlet, WTTV. The televising was accomplished with one RCA camera located in our auditorium balcony where the cameraman was able to traverse the entire stage. The image orthicon tube used required no extra light or makeup. The camera

was connected to the transmitter by a coaxial cable—the same one used to carry our athletic events. The cable has been permanently installed and is being used to carry special programs in addition to a regular weekly telecast produced by students of the radio classes.

"Technically these telecasts are proving to be no problem at all to school officials or to WTTV engineers." Wm. T. Voorhies

Dir., Radio & Dramatics Bloomington High School Bloomington, Indiana

ANOTHER PLEA FOR HELP

••• HAVE read with interest Donald Chadwick's letter in the December issue.

"I, too, am working for my ham ticket and looking forward to the day when I shall be able to pump some r.f. up the feeder lines, but like Donald, do not seem able to go past the 8-10 w.p.m. level. Anyway, on thinking of the many other fellows who have, I am encouraged to continue fighting. I am 20½ years old and shall be pleased to correspond with other prospective hams in any part of the world, 'CQ Jamaica.'

"Your magazine is always interesting to me, especially the 'Letters from our Readers' page and I'll close by saying 'Keep up the good work'."

Garnett W. Green 40 Arnold Road Cross Roads P.O. Jamaica, B.W.I.

NO SPACE

N THE March issue I missed the 'Letters from Our Readers' department. Could I have been so blind as to overlook it? Or can it be that the readers have not been writing? In case it is the latter, I am sending along my letter for next month's column as I enjoy the department very much.

"I get a big kick out of reading 'Mac's Radio Service Shop' by John T. Frye. As well as pointing out some very good ideas about the radio service shop he also adds a lot of humor to his story which makes it all the more interesting.

"I also like the editorial in the March issue warning people about the care needed in handling television picture tubes and about the high voltages involved in television.

"I also found in the March issue (at the bottom of page 158) a little item on checking voice coils that I would like to comment on. I would suggest to the writer to first place the ohmmeter test leads across the primary circuit of the output transformer. If the click is then heard it would indicate that the primary and the secondary circuits of the output transformer are OK as well as the voice coil circuit. This way all three circuits can be checked in one operation!"





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DOCKET 9295

A MEETING of the Federal Communications Commission held March 21, 1950 discussed the various developments in connection with the proposed new amateur regulations. Comments filed by interested parties included re-quests for a formal hearing, oral argument, and a poll of amateur licensees. The Commission decided that a formal hearing would not be warranted in view of the comments filed. It appeared to the Commission that little evidence would be adduced at such a hearing to aid the Commission in resolving the issues at stake. Accordingly the re-quests for a formal hearing were denied along with the request for a poll of the amateur licensees.

A general oral argument on the subject of the proposed rules has been set for May 19, 1950 in Washington, D.C. All interested parties who desire to participate in the General Oral Argument should notify the Commission in writing to that effect not later than May 10, 1950. The notification should consist of an original and three copies and refer to Docket 9295. The specific sections of the proposed rules to be discussed should be mentioned, together with an estimate of the time required for the presentation. -30-

S.T.E.N. HAMFEST

THE annual convention of the South Texas Emergency Net will be held

in Cuero, Texas, on May 27th and 28th. The event is a real hamfest with emphasis on emergency and net opera-tion. All amateurs, XYL's, and interested parties are invited to attend. There will be a transmitter hunt as

Part of the program. Registration will begin at Legion Hall in Cuero at 9 a.m. and the program will begin at 1 p.m. On both days there will be entertainment as well as technical information interspersed with contests. The YL's and XYL's will have special programs arranged for them.

Tickets may be obtained in advance from Robert Crossman, W5OUA, Box 214, Rockport, Texas. -30-

ERRATUM

The following resistor values were inadvertently omitted from the parts list accompanying the circuit diagram for the article "A Versatile Recording and Playback Amplifier" by Glen Southworth appearing in the March issue of the magazine. R_{28} is a 5000 ohm, 10 w. unit, R_{20} is a 50,000 ohm, 2 w. resistor, and R_{30} is a 50,000 ohm, 2 w. unit.

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May, 1950



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