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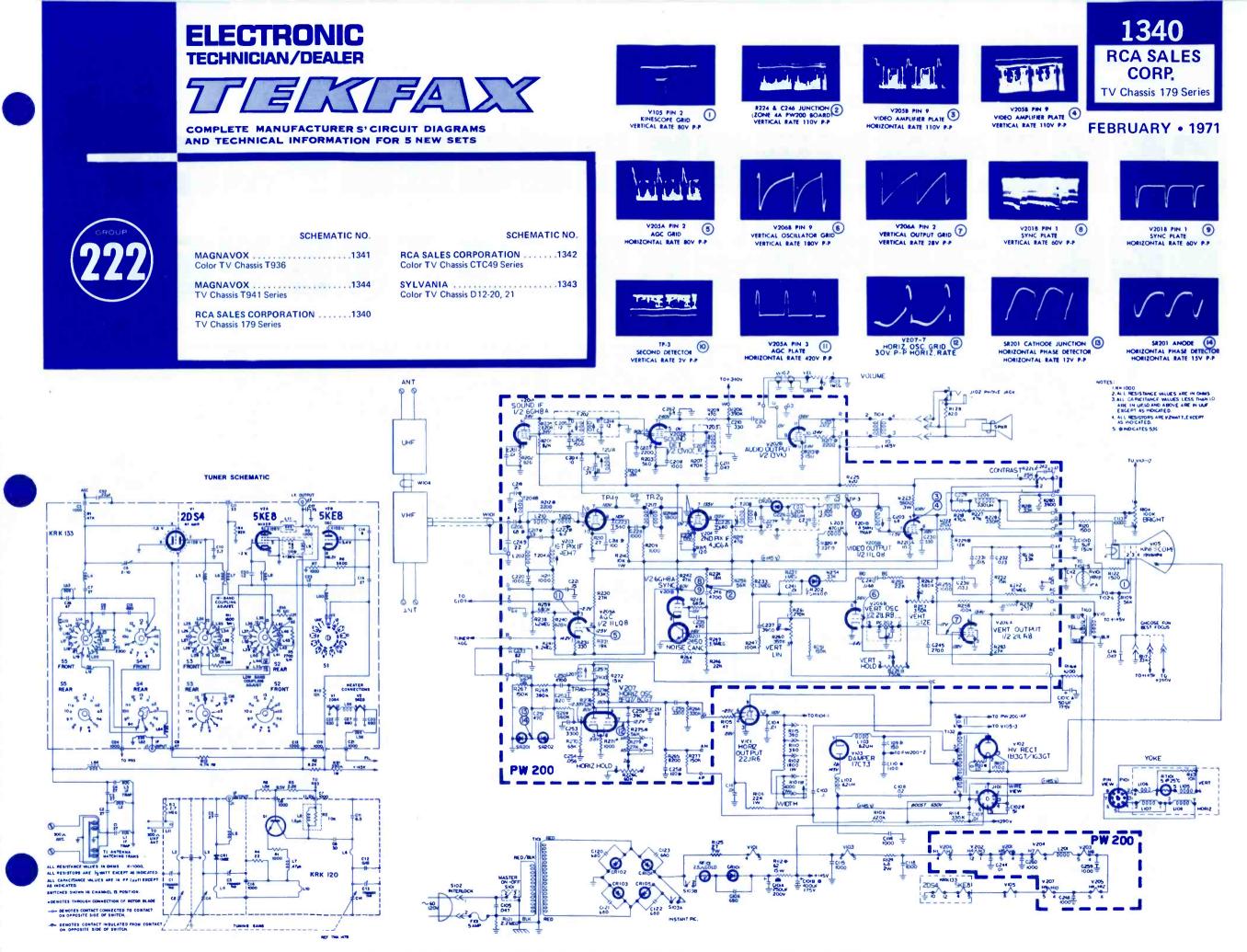
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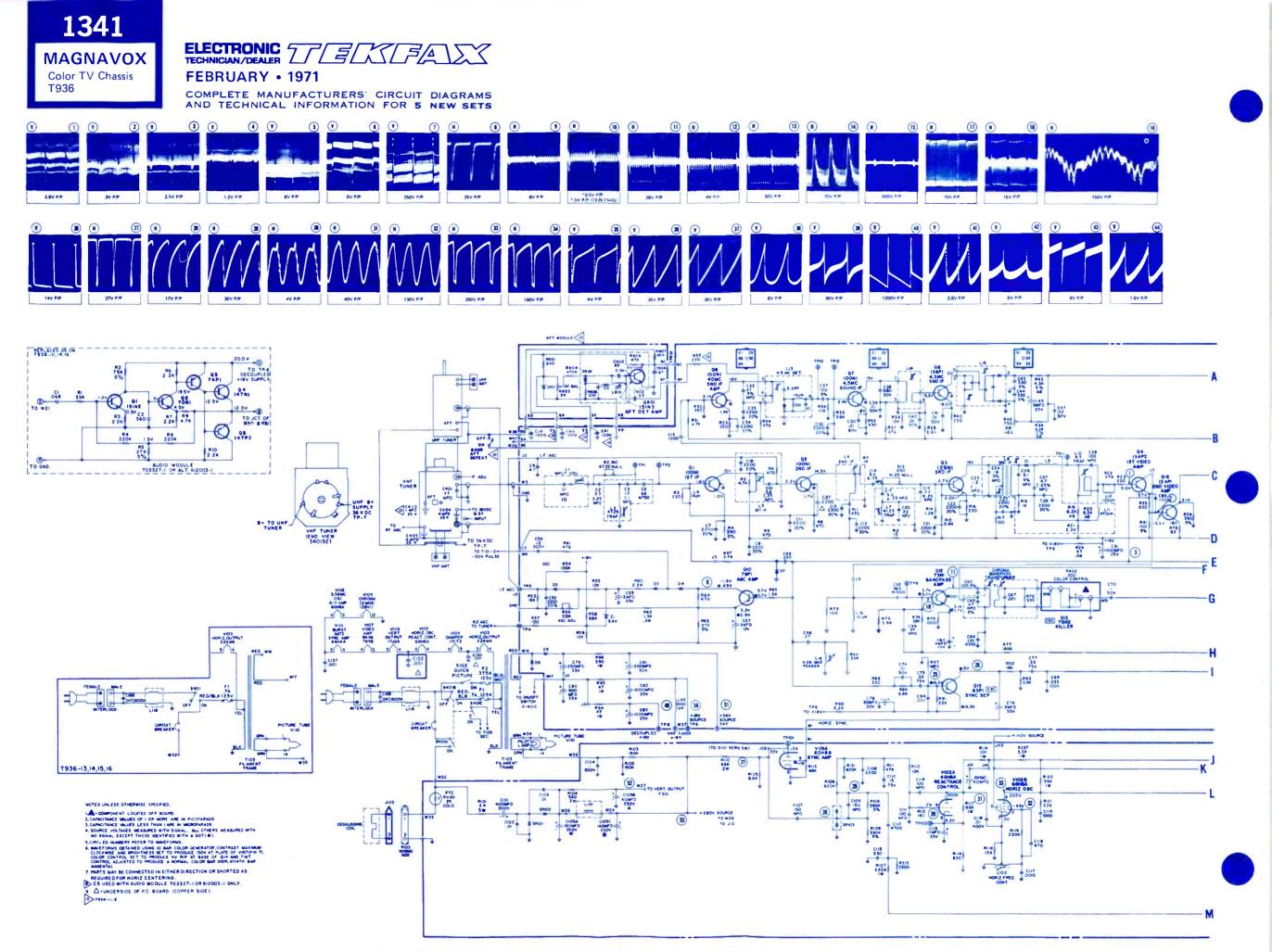
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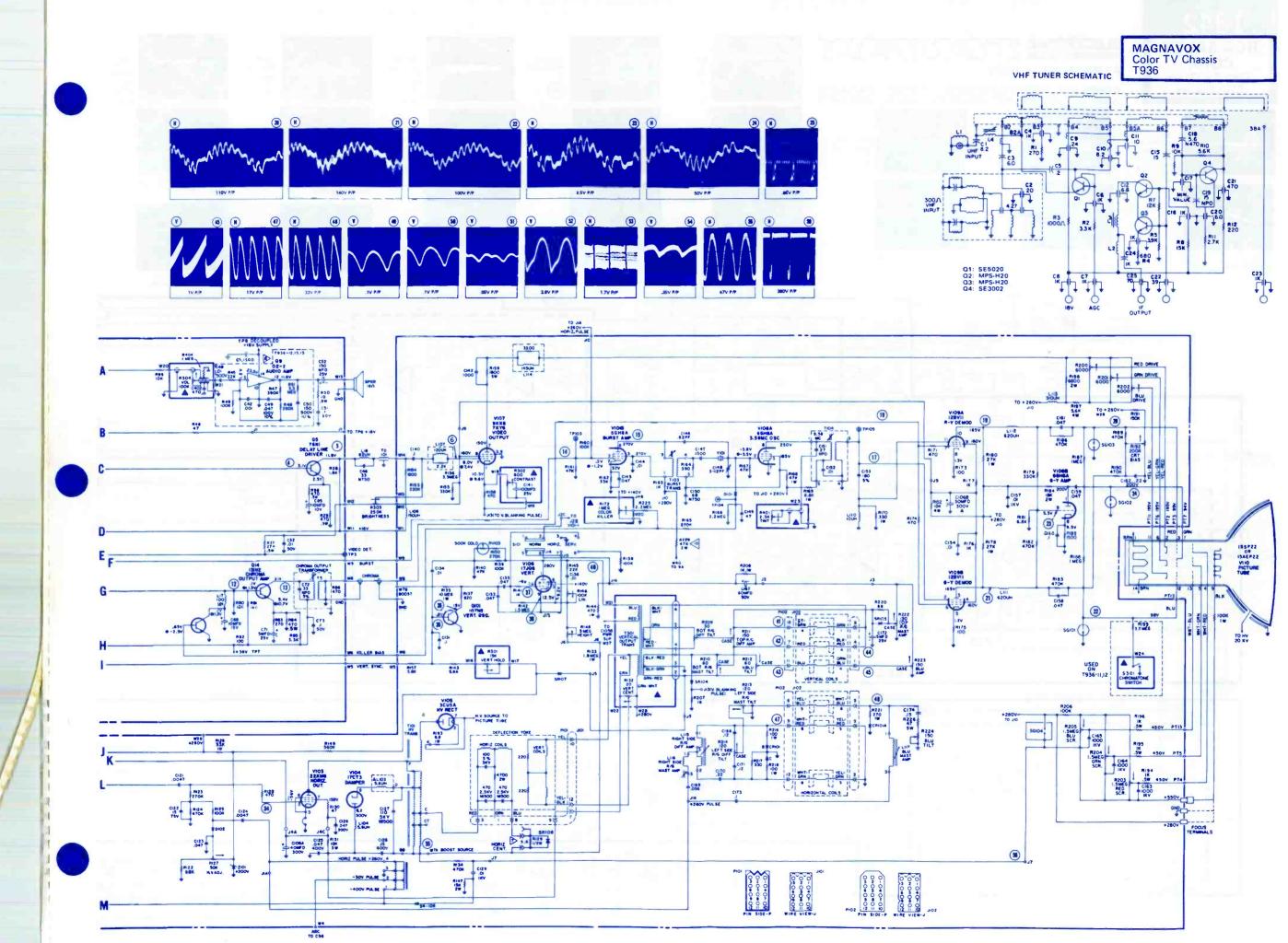
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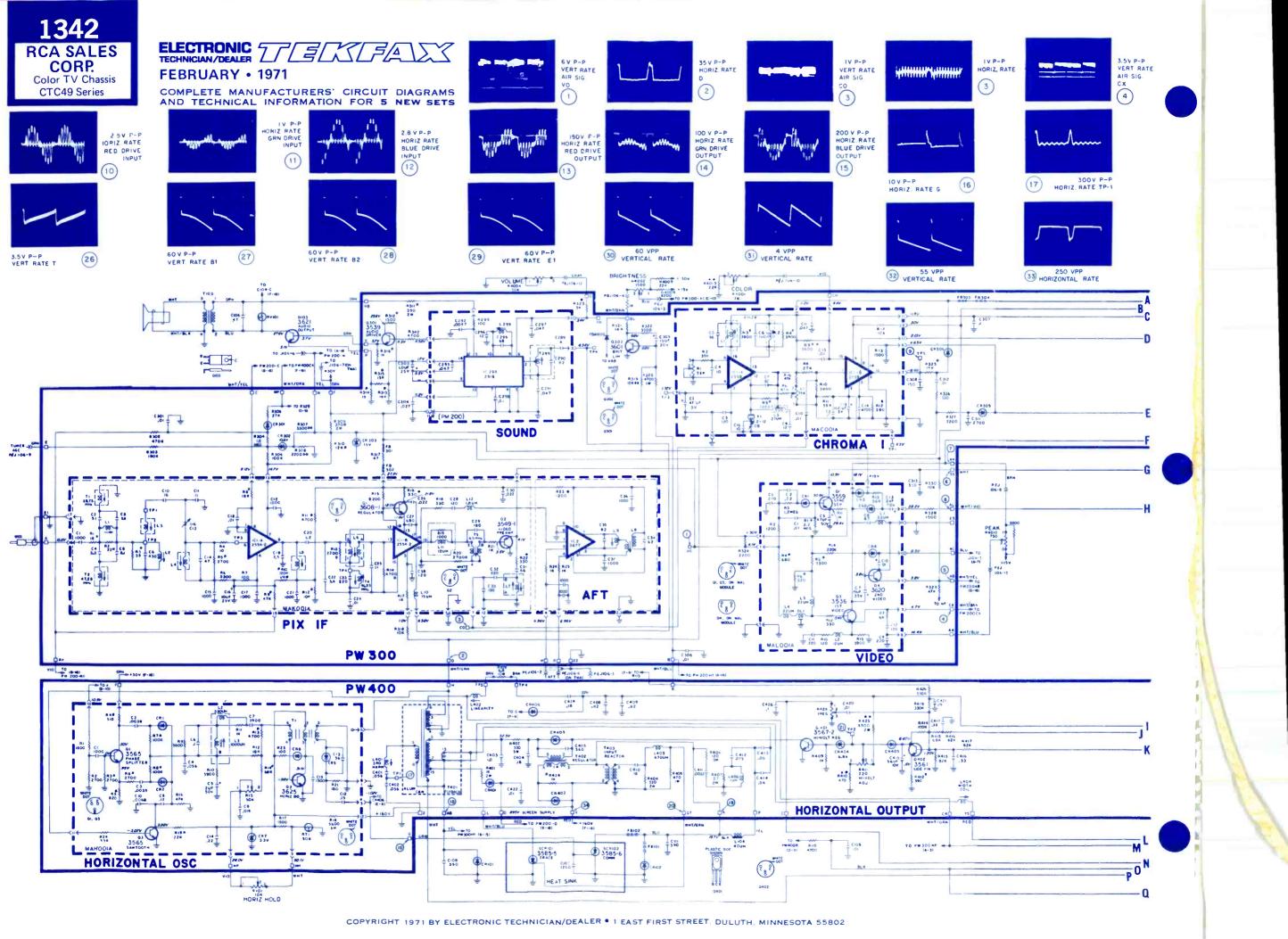
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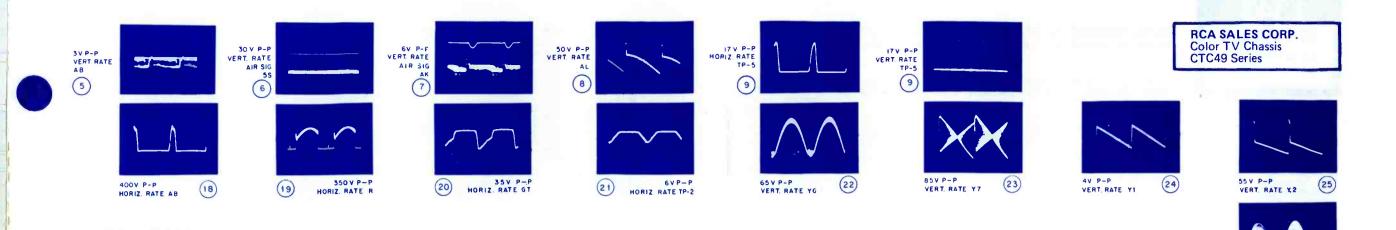


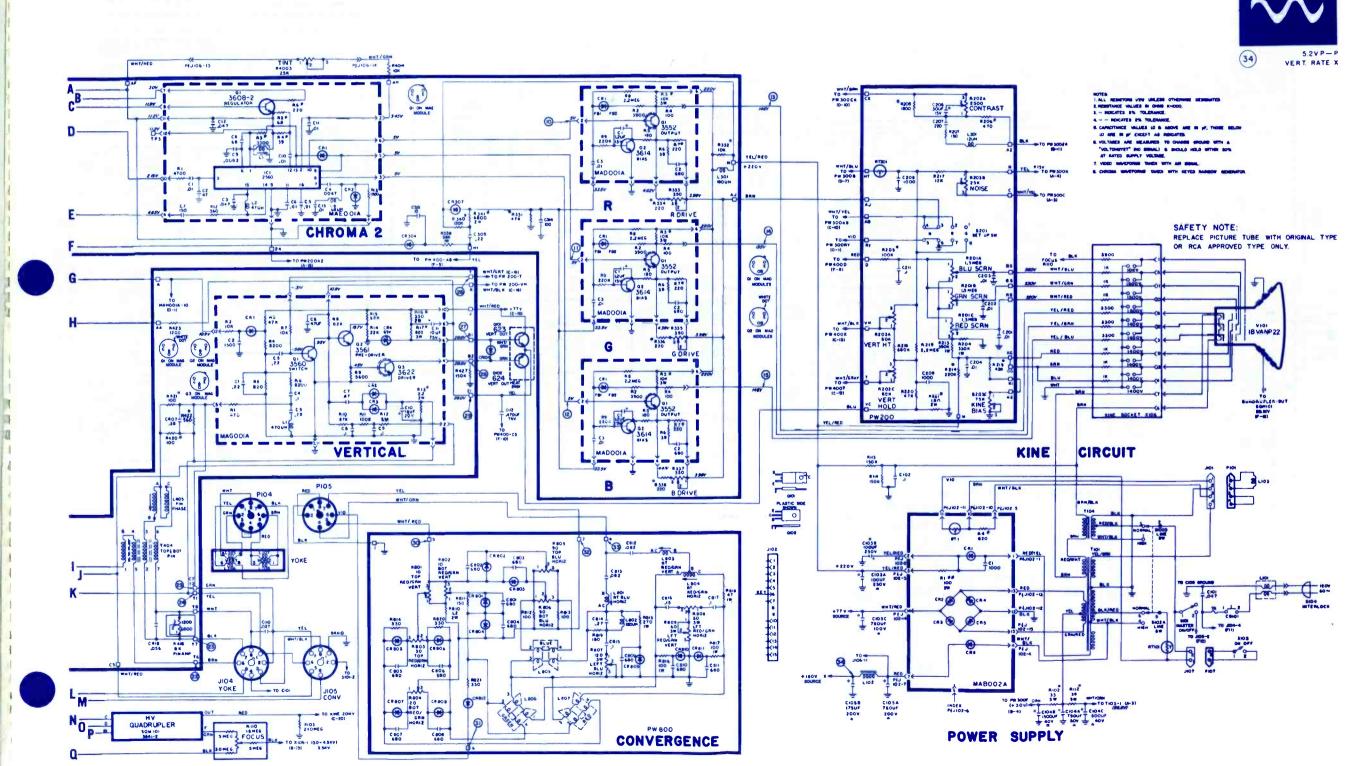




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ELECTRONIC TECHNICIAN/DEALER

COMPLETE MANUFACTURERS' CIRCUIT DIAGRAMS AND TECHNICAL INFORMATION FOR 5 NEW SETS

FEBRUARY • 1971

FEDRUART • 1	C522A~1000/40v elect		
SYMBOL DESCRIPTION	SYLVANIA PART NO.	C522B100/40v elect C522C50/400v elect	
		R124-4.5K · 15w	
C336A-10/400v elect		R126-VDR, 1ma, 76v	
C3368-50/150v elect		R274-5.6K - 4w	
C518A-100/400v elect		R444-13K - 7w	
C518B-100/400v elect		R514-VDR, 67ma, 20vdc	
C518C-20/400v elect	.41-27418-1	R516-thermistor · 1250, cold	
C518D-20/400v elect		L102-sound input	

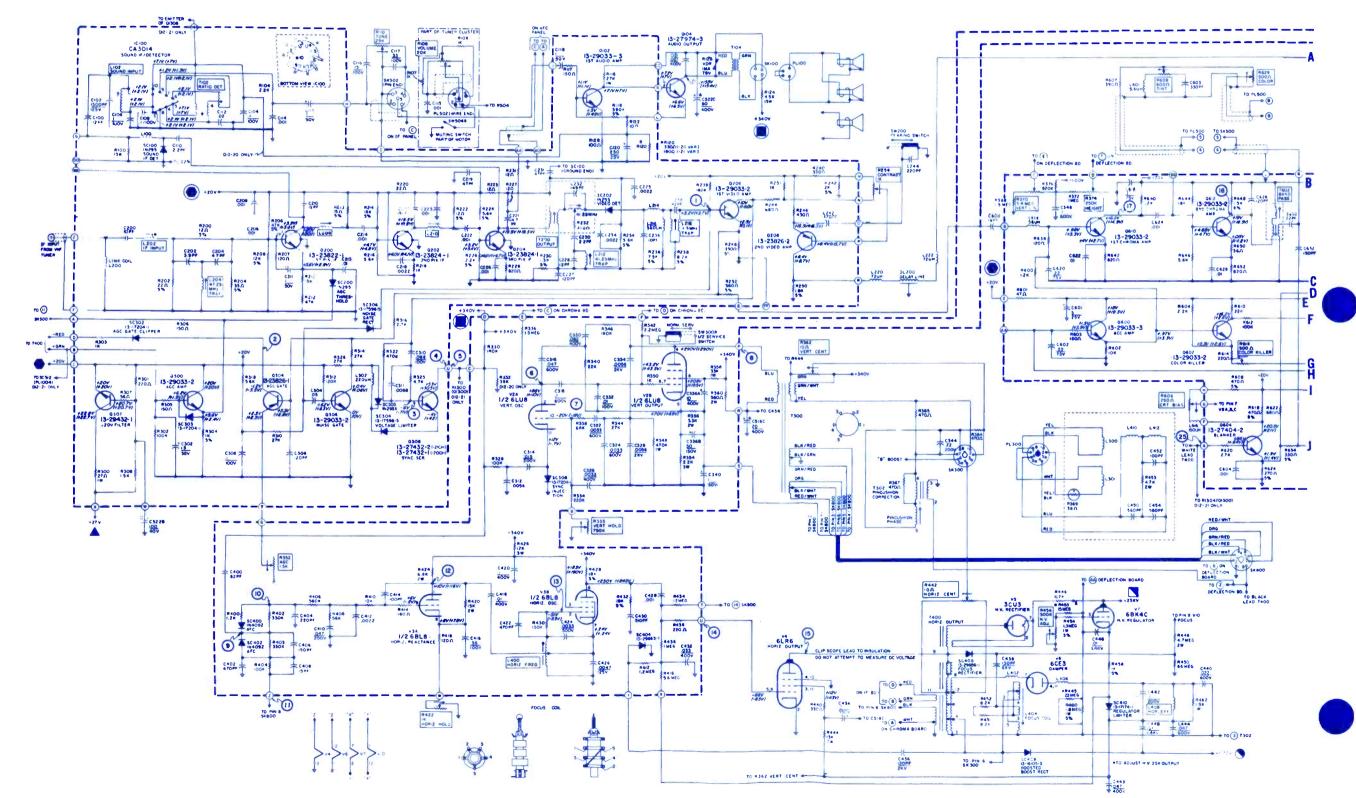
E202 WW. Stiller II Inport	
L204-47.25MHz trap	
L218-4.5MHz trap	
L400-horiz frequency	
L404-focus coil	
L408-horiz efficiency	
L600-5.6 µh	.50-15904-4
L610-reactance coil	
T102-ratio detector	
T104-audio output	
T200-44.5MHz output	
T300-vert output	.56-17559-6
T400-horiz output	.50-27415-1
T500-power xformer	.55-27416-3
T600-chroma output	
T602-band pass	
T604-burst	
SC100-IF sound detector - 1N295	
SC400,SC402-horiz AFC 1N4092	
SC406-focus rectifier	
R110-25K tone	

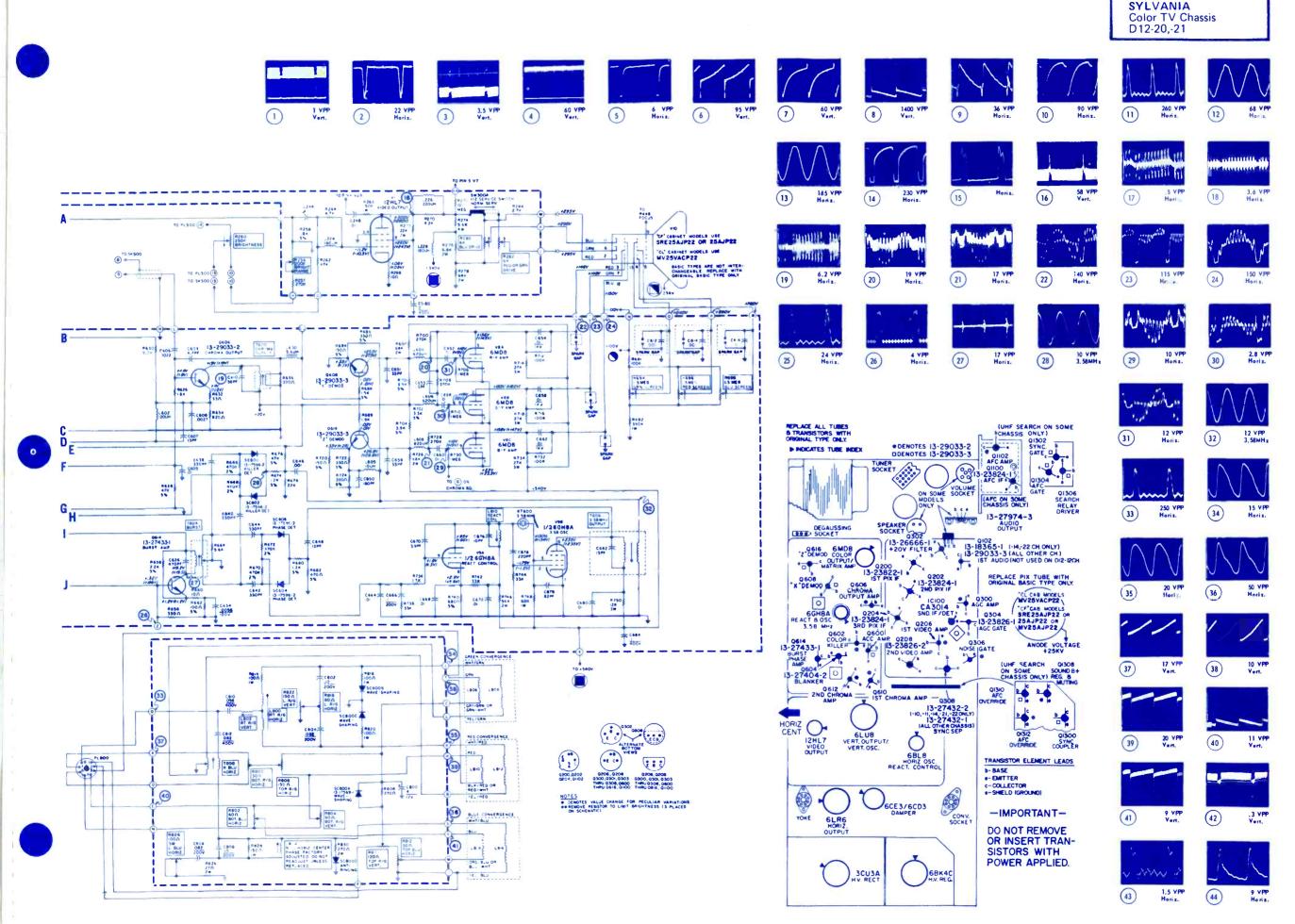
57-23827-1

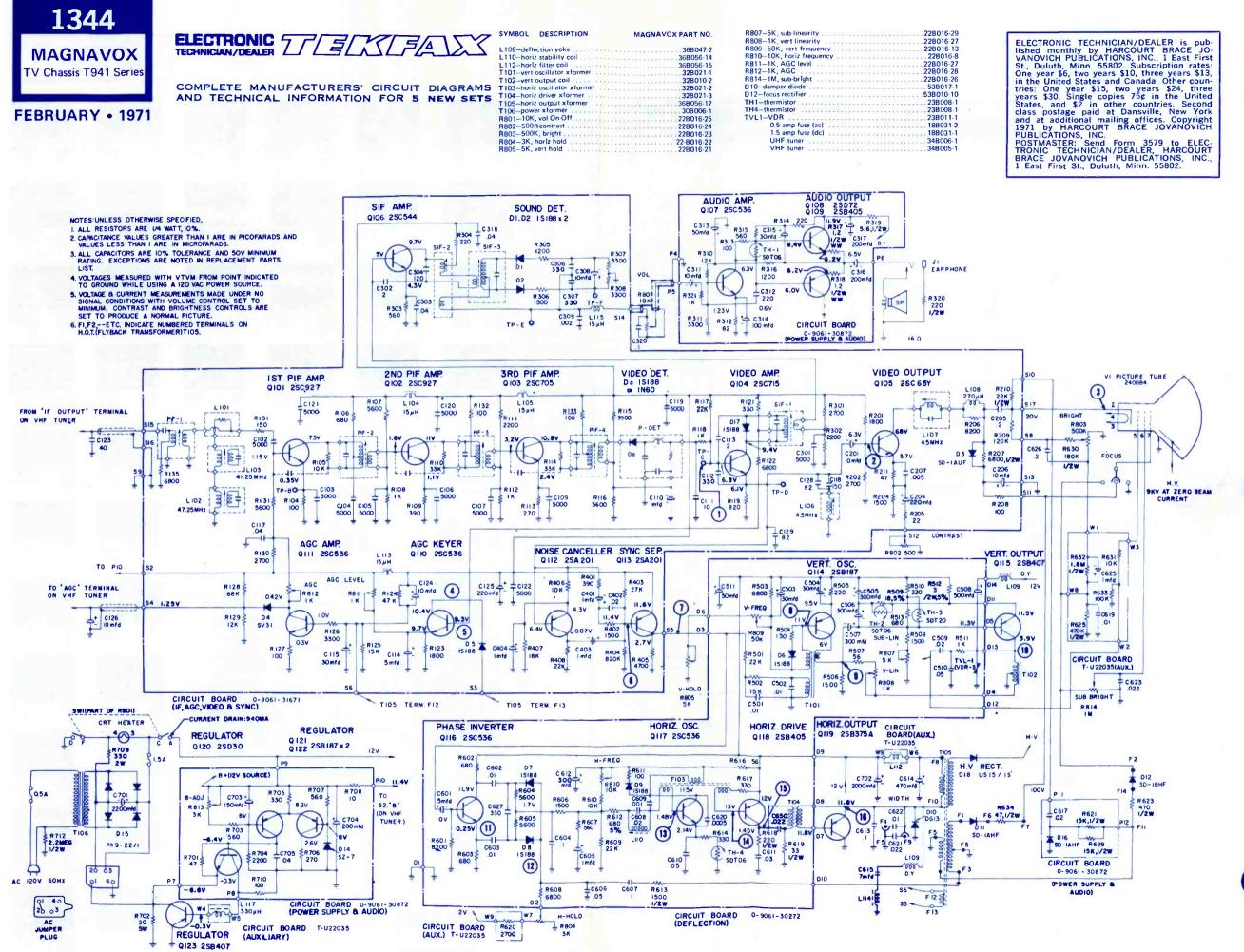
L202-44.5MHz IF input

.36-92898-51 .38-15257-18 .35-92495-71 .35-92495-54 .38-17072-1 .38-17071-1 .57-23832-2

R254–1K contrast	
R260-250K bright (-20)	
R260-250K bright (-21)	
R280-6K blue drive	37-11632-18
R335-750K vert hold	
R352-1.5K AGC	
R362-100 vert center	
R370–3.4M vert Ilnearity	
R422-1K horlz hold	
R442-100horiz center	
R456-500K HV	
R606-250 CRT blas	
R609-600 tint (-21)	
R616-5000 color killer	
R629-5000 color (-20)	
R629-500n color (-21)	
CB500-circuit breaker	
tuner · UHF	
tuner VHF (-20)	
tuner VHF (-21)	
yoke - deflection	







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FEBRUARY 1971 . VOLUME 93 NUMBER 2

This month's cover, supplied through the courtesy of P. R. Mallory Co., depicts a burglary in the process of being attempted. It may be thwarted if the house is equipped with an ultrasonic intrusion alarm.

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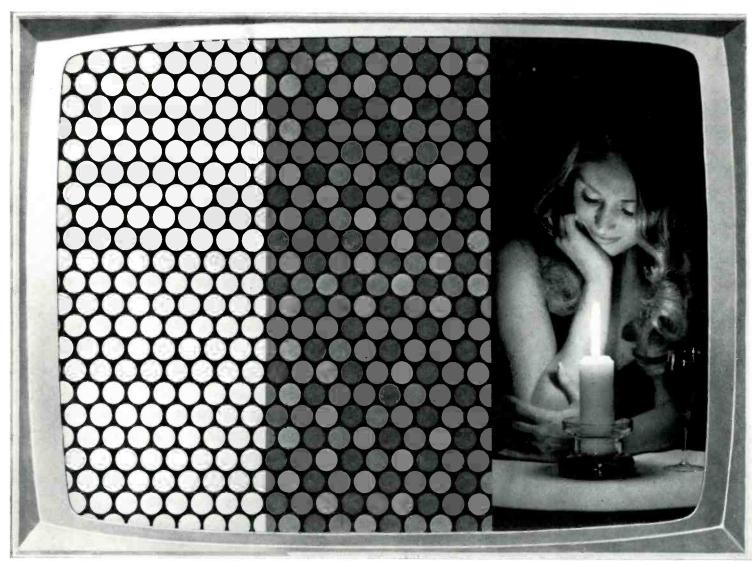
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EDITORIAL

Industry Involvement



ELECTRONIC TECHNICIAN/DEALER exists to serve you as a member of the electronics industry. Although Joe Zauhar's and my personal electronics backgrounds, plus our continual personal contact with others in the electronics industry, helps us maintain an awareness of your interests and needs, **that is not enough!**

You, our readers, have fortunately been a very vocal group, frequent to express your needs, interests and opinions by either letter or telephone. When comparing notes with other editors working in the same building, it becomes apparent that our industry is unique in that respect. Although as dealers and technicians we find ourselves placed in competitive positions, it is not the kind of competition encountered elsewhere. In other industries, the extreme jealousy over trade secrets often results in communication barriers. We, however, can be proud of the fact that those in our industry are typically congenial, having the "gift to gab," the desire to get together to "cry in each other's coffee," and warm hearted enough to care about the other fellow. In what other industry would a Readers' Aid Column—such as the one published in ELECTRONIC TECHNICIAN/DEALER—result in such good responses, promptly providing those asking for help with assistance from their fellow readers? Many a reader has sent us a note expressing surprise at the amount of help received from other readers as a result of our publishing his letter.

We appreciate your active interest in ELECTRONIC TECHNICIAN/DEALER and look forward to continuing such communication. We intend to continue to respond to your comments so that this publication will fulfill your professional needs.

It is our belief that with the increasingly complex technology that must be mastered for effective sales and servicing, ELECTRONIC TECHNICIAN/DEALER must continue to emphasize that aspect of our industry—devoting its primary attention to that subject. However, it is not enough to merely possess adequate technical skills for mastering today's and tomorrow's electronic circuitry. We are in a field which requires continual contact with the public and a good business "sense." As you have probably noticed, we recently began including articles related to that subject.

Despite the outgoing nature of most technicians and dealers in our industry, we are inclined to be quite independent in our thinking and actions. Only when faced with some common threat do we tend to ban together. Although we do care about each other, as proven by our Readers' Aid Column, the professional associations across the country represent only a small portion of our readership. We feel that this is unfortunate since these local, state and national associations are excellent for supplementing the technical and business information included in our publication, they keep us abreast of service charges prevalent in the region, help us to compare wholesale prices, make us aware of outside threats to our business and help us in petitioning for local ordinances which may offer us better protection. Some associations even offer credit plans to help improve retail sales and group insurance protection.

As a publication, we are also very interested in these associations. Unfortunately we are generally advised of meetings too late to publish that information or attend, but we are attempting to rectify this situation. We would appreciate it if you would advise us three months in advance concerning the date, location and activity planned for your association meetings. This will permit us to publish that information, thus hopefully improving industry interest and attendance. And if these meetings can be fitted into our schedule, we hope that it will be possible for us to be there with you.

Phillips Dahlen

LETTERS

I suppose it is the crust left over from Christmas celebration that gives me the nerve to comment on your excellent article, "Tuned Circuit Signals."

Fig. 7 causes me some wonder that there is no crossover point, which one would expect with a two-to-one frequency input [or a single curved line (complex or not) if in the correct phase—edge on]. I also tend to think that the pictures over Fig. 19 and 20 should be exchanged.

To me, it would seem that some element of the fundamental square wave was present in the scope pictures with the exception of Fig. 19 (corrected). Since "continuous sweep" produces an even sinewave and all the inputs using triggered sweep or even the other Lissajoux patterns all show something akin to Decay—the response to a step function, I speculate that there is some unwanted pick-up.

Fourier would lead us to believe that each of the harmonics of the wave should be of constant amplitude, whatever their phase relationship to the fundamental. However, in all cases

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but one there is a pattern that decays over the period of a half wave, or between the positive and negative excursion (step) of the fundamental. In my humble opinion, if the fundamental was truly filtered out, there would be some sort of pattern that would have continuity over the full period of the fundamental and not such definite "separate halves."

You state that the horizontal input in most of the illustrations was the fundamental sinewave from the generator and that there is a noticeable decay factor in each figure. But then for Fig. 21 you state that you are using the scope's sinewave sweep and I see that there is *no* noticeable decay—instead it is the pattern that would be expected when beating two sinewave frequencies in the scope (Lissajoux). It must therefore be concluded that the sinewave output of your signal generator includes higher partials and is not a sinewave at all.

Believe me, I am all for this sort of article and would not want my comments to discourage you from suggesting that such investigations are not just textbook fillers, but have actual practical value and can be reproduced on any workbench. I think it was unfortunate that you did not mention that Fig. 22 represents a 90° phase difference between the horizontal and vertical inputs and that the rotating, slanting ovals represent other phases in addition to not exactly matched inputs (frequency). These are all things that help to get the most out of a scope. Many technicians are all too prone to throw everything aside that smacks of theory, saying: "We don't have to know about that sort of thing."

JOSEPH G. BRADLEY, JR.

We appreciate receiving your detailed letter. It is good to know that after going to the effort of digging out information for an original article there are subscribers, such as you, who are interested enough to read the material closely.

You are correct in stating that the photographs for Fig. 19 and 20 were transposed on page 51 of our December 1970 issue.

In the more conventional scope trace (Fig. 18) we see that the harmonic frequencies are subject to decay. This is not what would be expected from the Fourier mathematic model of square waves. However, this theory ignores one important reality—it assumes that all components are without internal resistance and function at 100 percent efficiency. The rate of decay observed is dependent on the tuned circuit's efficiency or Q factor. The higher the Q factor, the lower the rate of decay. By using two resonant circuits tuned to the same resonant frequency, we can increase the total effective Q factor and reduce the rate of decay to that shown in Fig. 20 (corrected) of the article.

One simple analogy can help further demonstrate the limitations of the Fourier concept of square waves. A square wave can be compared to a voltage that is switched ON and OFF at a constant rate. But assume that this rate is but once every 2 sec. We can not expect the surge in current, which occurs as the circuit is switched, to be sufficient to maintain a continuous harmonic sinewave during the entire 2 sec. interval.

You are correct in assuming that Fig. 7 should appear looped like "a figure eight" or a diagonal line if sinewave signals are applied to both the scope's horizontal and vertical inputs and one signal is twice the frequency of the other. However, since we are instead comparing a decaying sinewave with a sinewave, the waveform shown in Fig. 7 results.

Although not very noticeable, some decay is apparent in Fig. 21, just as it is in Fig. 20 (corrected). The difficulty in observing the decay results from the increased Q factor of the dual-tuned circuit and not a distortion of the signal generator's sinewave output. Although not shown in the article, the EICO signal generator has a well formed sinewave output.

Signals of the same frequency were applied to the scope's horizontal and vertical inputs for producing the pattern shown in Fig. 22, 23, and 24. You are correct in indicating that Fig. 22 shows a 90° phase relationship while the slanting ovals represent other phase angles. These phase angles are also covered in this month's article, "FM Stereo Alignment," which begins on page 38 and will be covered in future issues. Ed.



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I am looking for the schematic of an organ manufactured by the Minshall Estey Organ, Inc., Brattleboro, Vt., Organ Model H. I was informed that the company had moved by the returned letter that I had mailed them. Does anyone have any information concerning where I may purchase the schematic?

GEORGE OLSEN

George Olsen TV 13519 Westwind Drive Silver Spring, Md. 20904

I have for sale a set of John F. Riders Radio Perpetual Trouble Shooter's Manuals, Vol. I through Vol. XV in good condition. I will sell these by the set or by each volume at \$3.50. RUSSELL G. SPLAR

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R. IWASYK

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I am in need of the schematic and instruction book for the TV field strength meter, Model A-460, made by Approved Electronics Instrument Corp., New York, N.Y. I understand this company is not in business anymore.

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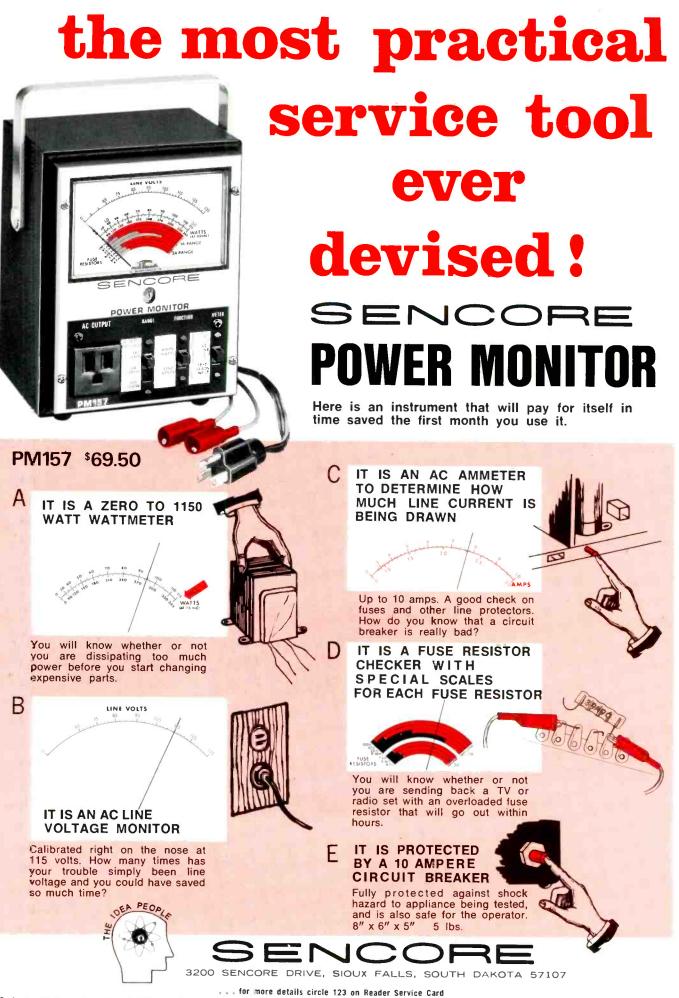


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NEW AND NOTEWORTHY

For additional information on products described in this section, circle the numbers on Reader Service Card. Requests will be handled promptly.



OSCILLOSCOPE 700

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A Model 556A Oscilloscope is designed so that the controls are grouped on the front panel by function for more convenient use. The vertical amplifier is reportedly ac/dc coupled and fully compensated for optimum response with a sensitivity of 20mv/cm over a dc to 1.5MHz bandwidth. Specifications indicate that the instrument has a sweep frequency of from 10Hz to 100kHz in four ranges and is continuously variable in the respective ranges. The scope features solid-state circuitry and a compact chassis measuring 6¹/₄ in. W by 10¹/₂ in. H by 17³/₄ in. D and weighing 18³/₄ lb. Price \$239. Kikusui Electronics.



TAPE DECK 701

Features three motors and three heads for quality recordings

The three-motor, three-head stereo tape deck, Model 640, is said to incorporate many of the features found in the more expensive models. These reportedly include a record equalization selector switch, which optimizes performance with both standard tape and low-noise high-output tape; a die-cast tape guide and head black mounting frame for permanent alignment of critical transport components; plus front panel sound-on-sound, echo controls, and microphone and line mixing. The tape deck is said to also feature mechanical memory capability, which permits timeractivated recording, playback and shut-off, and positive-acting lever-type transport controls. Price \$369.95. Superscope.



AM/FM TUNER 702

Slips into all automotive 4and 8-track cartridge players

A new cartridge AM/FM and FM-multiplex tuner, Model 30-3075 has been designed for automotive stereo tape players. The unit will reportedly slip into the cartridge loading slot of all automotive 4- and 8-track tape cartridge players. It is powered by an internal 9v battery and slides into the cartridge slot, after being connected to the car's external radio antenna. A special antenna adapter connector is said to be included, which can be permanently left in the car's antenna line. A slide switch on top of the unit selects AM or FM modes. Price \$45, GC Electronics.

FOR MORE NEW PRODUCTS SEE PAGE 56

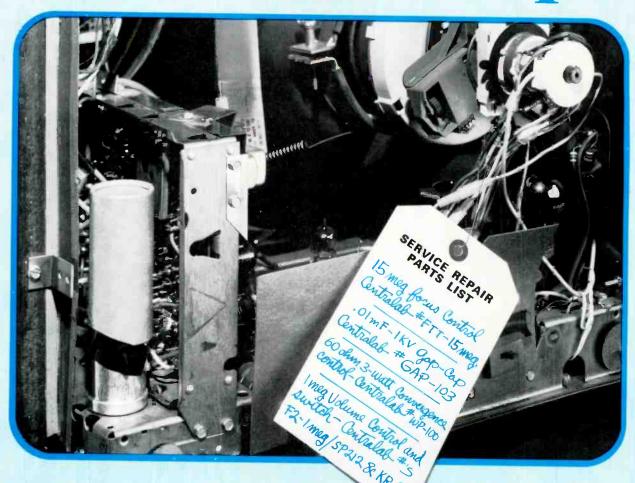
UHF VERY WEAK	UHF WEAK	UHF STRONG	NO UHF			
CS-U3 \$26.40	CS-U2 \$18.35	CS-U1 \$12.05	All Prices Subject to Change	NO VHE		
CS-U3 \$26.40 NOTE: Each model above is also available with 75 ohm coastal cable downlead and compact behind-the-set balun-splitter to match and	70-12B \$44.95	70-12A \$39.95	70-V11 \$26.95	VHF STRONG		
70-18B \$62.95	70-18B \$62.95	70-18A \$56.95	70-V14 \$34.95 70-V17 \$39.95	VHF MODERATE		
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70-29B \$100.00	70-29B \$100.00	70-23A \$73.95	70-V28 \$74.95 70-V35 \$94.95	SPECTRUN		
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ELECTRONIC TECHNICIAN/DEALER FEBRUARY 1971

TEKLAB REPORT

The receiver's vertical feedback circuit, known as a Miller run-down circuit, was seldom employed before the advent of transistors, due to its high gain requirements

Introducing RCA's Argosy Portable Color - TV Set - Part II

■ Most of the repairs that might eventually be required for this chassis can be accomplished by module replacement rather than by replacing single defective components. However, it is still important to know how each circuit functions for more effective servicing or possible module repair.

Last month's Teklab Report reviewed the RF, IF, AGC, Noise Control, Video and Kine Driver circuits, while this month we cover the power supply, and horizontal and vertical deflection circuits, which in many ways are different from comparable circuits in earlier RCA receivers. The complete schematic for this RCA CTC49 chassis can be found in the February Tekfax, Schematic No. 1342.

Power Supply

Although the power supply utilizes a transformer, one side of the ac line is connected to the chassis. Its dc outputs and their principal uses are as follows:

- The 220v source powers the kine drivers, Modules MAD. A halfwave rectifier and an RC-pi circuit are used in supplying this filtered voltage.
- Four diodes in a bridge configuration provide two additional power sources. One of these, us-

ing capacitive filtering, supplies about 77v to the vertical-output transistors. The second is divided into separate supplies—one for the low-level transistors used throughout the receiver, the other for the audio system. Both are nominally 30v sources and both use RC-pi filters.

• A half-wave rectifier, with an LC-pi filter, is used to supply 160v to the horizontal deflection system.

As illustrated in the schematic diagram, Fig. 1, the receiver contains a rather unusual degaussing circuit. In this schematic, T101 is the power transformer (for simplicity several windings were deleted from the illustration) and S102 is the NORMAL/ HIGH line switch. At "turn-on" the resistance of temperature compensating thermistor RT1 is low and the degaussing current is high. As thermistor RT1 warms, its resist-

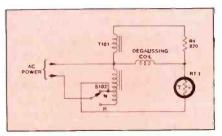
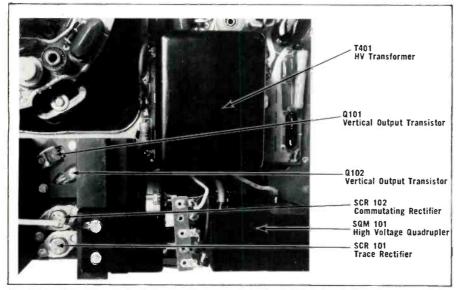


Fig. 1—Simplified schematic of the automatic degaussing circuit.

ance increases until the degaussing current approaches zero. After warm-up, the voltage drops across resistor R4 and thermistor RT1 are equal to the voltage drops across the upper and lower transformer windings, respectively; making the voltage across the degaussing coil equal Ov. The current, which still flows through resistor R4 and thermistor RT1, keeps the latter warm to maintain its high resistance.

Horizontal Deflection Circuits

The CTC49 chassis employs a horizontal deflection and high voltage system similar to the one employed in the CTC47 chassis, which was derived from the circuits in the CTC40 chassis. However, because of the wider CRT deflection angle used in the CTC49 chassis, a more sophisticated means of correction is necessary to overcome side pincushion, the horizontal scan must be increased when vertical scan is near the center of the raster. In earlier RCA color-TV sets, this correction was accomplished by passive components. Since the highvoltage regulator of this chassis also controls the scan width, side pincushioning may be corrected by providing a second input to the regulator. This input is derived from the vertical deflection circuits and proc-



Rear view of chassis showing components employed in the horizontal and vertical sweep circuits of the RCA CTC49 color chassis.

essed by the circuit shown in Fig. 2. To correct side pincushioning, it is only necessary to increase the forward bias of transistor Q402 when vertical scan is near the center of the raster and decrease it when the vertical scan is near the top and bottom. One output from the vertical deflection system is fed to the base of transistor Q402 by way of resistors R416 and R415 while another arrives through resistor R417. These samples of vertical deflection signal are shaped into a parabolic waveform which reaches its maximum positive potential at vertical midscan.

Two additional inputs from the horizontal deflection system are used to optimize this high-voltage regulation and pincushion correction. The first of these is obtained from terminal D (Fig. 3) of the high-voltage quadrupler through resistor R115, and reaches the base of transistor Q402 by way of resistors R426 and R419. It allows the regulator system to "measure" the beam current and more accurately regulate the high voltage.

The second input is obtained from terminal C of the quadrupler and reaches the base of transistor Q402 by way of potentiometer R428. This input samples the high voltage by means of the capacitive voltage divider, made up of capacitor C426 and the capacitors in the quadrupler. It compensates for phase shift of the side pincushion correction voltage, which is the result of the capacitance of the kinescope ultor (second anode) connection.

The amount of effect which this sample from the quadrupler will have on the pincushion amplifier may be adjusted with potentiometer R428. The amplifier has been designed so that when brightness is set for a barely visible raster and potentiometer R428 is set to minimum (CCW) there will be no pincushion. Potentiometer R428 is then adjusted to correct the pincushioning, which will appear when the brightness is increased to maximum.

Unlike conventional high-voltage power supplies, which rectify a positive pulse from the flyback transformer with a half-wave rectifier, the CTC49 chassis uses a solid-state quadrupler to produce high voltage. This reduces the required pulse amplitude from about 23kv to normally 6kv. The quadrupler itself is hermetically sealed and is not repairable.

Vertical Deflection Circuits

The vertical feedback circuit employed in the CTC49 chassis is known as a "Miller run-down circuit." This circuit was seldom used until the advent of transistors because of its high gain requirements. The Miller run-down circuit multiplies the changing capacitance by a factor equal to the gain of the amplifiers without feedback, resulting in a very linear output. Many of the variations in supply voltage, amplifier gain, etc., which would drastically change the output of a conventional vertical-deflection circuit, would have very little effect in the Miller circuit because of the large degenerative feedback.

A schematic of the RCA CTC49 chassis vertical-output stage is shown in Fig. 4 with a simplified yoke circuit. The circuit configuration is similar to a high-quality audio amplifier. The yoke itself is analogous to a speaker voice coil, capacitor C419 to the coupling capacitor, and R_x equivalent to the total resistance of the yoke and convergence circuits. The value of capacitor C419 has been chosen to provide maximum energy transfer at the vertical scanning frequency. Feedback to the Miller capacitor is developed across resistor R13, and capacitor C10 is a filter.

During retrace, transistor Q3 is cut off, allowing its collector voltage to increase towards B+. The 65v zener diode, CR4, limits the maximum base bias of transistor Q101, serving to limit yoke retrace current. During scanning time, the bases of transistors Q101 and Q102 are driven progressively less positive at a linear rate. Conduction is through transistor Q101 during most of the retrace time and as the scan passes from the top to the center of the raster. The voltage across capacitor C419 reaches its maximum at the center of vertical scan (90° out of phase with the current), and during the lower half of the scan capacitor C419 discharges back through the yoke and transistor Q102. This current increases at a linear rate, since the forward bias on the base of transistor Q102 is also increasing at a linear rate.

The diode connected between the bases of transistors Q101 and Q102 improves the switching characteristics of the transistors at mid-scan. Transistor Q102 has no bias as long as transistor Q101 is conducting. Therefore, only slight voltage swings are necessary to cut off transistor Q101 and turn on transistor Q102 at the center of the raster. If the diode were shorted or bypassed, reverse bias would exist between the base and emitter of transistor Q102 while transistor Q101 was conduct-

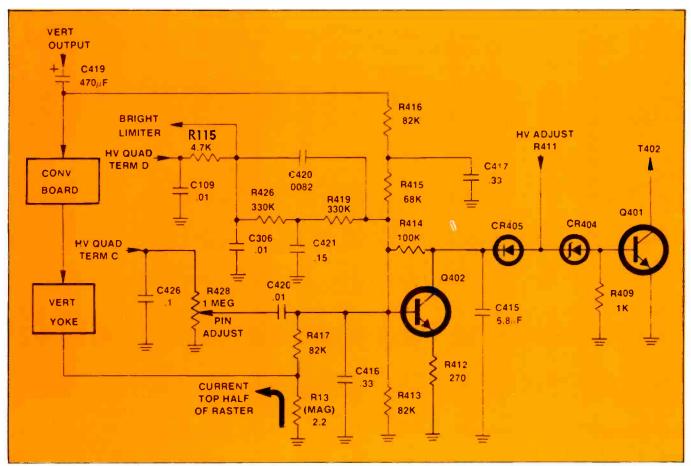
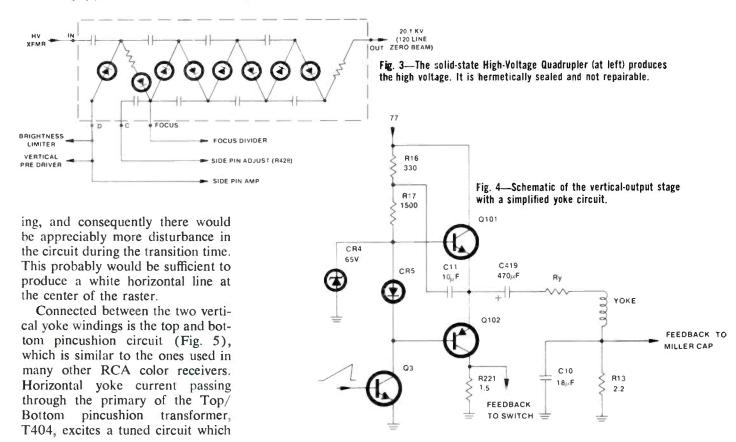


Fig. 2—Diagram of the side pincushion amplifier circuit. The high voltage regulator also controls the scan width. The side pincushioning is corrected conveniently by providing a second input to the regulator.



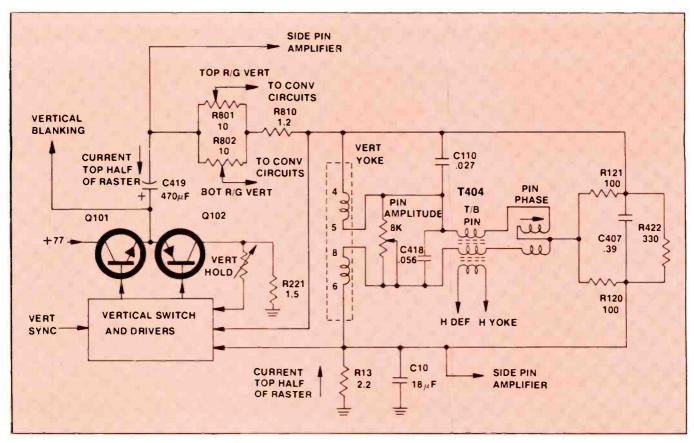


Fig. 5-Yoke circuit of the RCA CTC49 color chassis.

consists of the pincushion phase coil and capacitor C418, which is in series with the vertical yoke.

Depending on its phase, this 15.73kHz voltage may either buck or boost the vertical deflection. The core in the pincushion coil is adjusted so that vertical-yoke current is decreased at the edges of the raster and increased at the vertical centerline. The 8K potentiometer controls the amount of yoke current buck and boost, so that the inherent pincushioning is cancelled.

Passing through the second yoke winding, terminals 8 to 6, the yoke current reaches ground through resistor R13. The voltage at the junction of resistor R13 and the yoke is positive during the top half of scan and negative during the bottom half. This voltage is fed to the vertical amplifier through the Miller feedback loop and also to the side pincushion amplifier, where it is used for wave shaping.

The components of the predriver stage are physically located in a number of places in the instrument; locations may be determined by the symbol numbers. Component symbols of one or two digits (C9 and R11, for instance) are in Module MAG. The 200-, 300-, and 400-series symbols indicate locations on boards PW 200, PW 300, and PW 400, respectively.

When comparing the vertical predriver circuit in Fig. 6 to the simplified circuit in Fig. 7, we see that transistors Q1 and Q2 have the same functions in each diagram, and capacitor C1 of Fig. 7 is analogous to capacitor C7 of Fig. 6. Resistor R2 of Fig. 7 is replaced by resistor R8, plus a number of other components also included between the base of transistor Q2 and B+.

In understanding the functions of these additional components, remember that anything which increases the positive voltage at the junction of resistors R8 and R203A will increase the raster height. Consider first the major voltage source to resistor R8, consisting of resistors R203A and R217. As the value of resistor R203A is decreased, height will increase. If the setup switch, S201, is closed to the service position, the voltage applied to resistor R8 is diminished to practically 0v, collapsing the raster.

Because transistor Q102 is cut off

while the top half of the raster is scanned, the voltage at its collector is 0v until the vertical scan reaches its center. During the bottom half of scan, the collector current of transistor Q102 increases linearly, so that the voltage fed back to resistor R8 tends to "stretch" the lower part of the raster, to overcome a tendency towards bottom compression. Feedback to the vertical-switch transistor also is derived from resistor R221.

The remaining input to resistor R8 comes from the horizontal system. As the high-voltage return current from the quadrupler to the brightness limiter increases or decreases, the voltage at the junction of resistors R426 and R320 also varies. An increase in kine current reduces the voltage to the height control, causing a slight decrease in vertical deflection. This causes scanning height to track scanning width.

The most important feedback signal in the system is fed to capacitor C7 from the junction of resistor R13 and the yoke. Ignoring the effect of capacitor C10, the voltage at this point reaches its maximum positive value at the beginning of scan, pass-

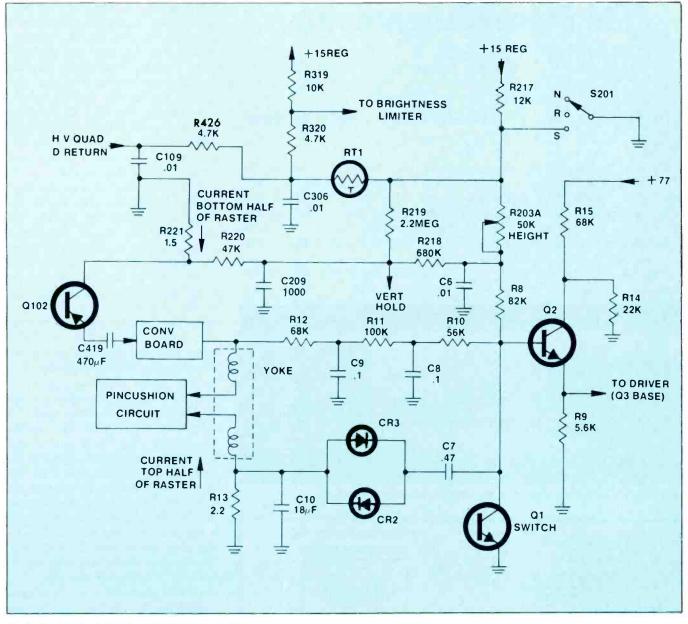


Fig. 6-The Vertical Predriver Circuit with its inputs.

es through 0v, and reaches its maximum negative value just before vertical retrace. Therefore, the feedback to transistor Q2 is degenerative, since the voltage at the base of transistor Q2 tends to rise throughout the scanning interval. Capacitor C10 is used to filter out any horizontaldeflection voltage present.

The switch transistor Q1, and its circuit (Fig. 7), perform three functions—it controls the free-running frequency of the vertical deflection system, allows synchronization with the received signal and determines the duration of vertical retrace.

Information used in this article is based on material supplied through the courtesy of RCA Sales Corp.

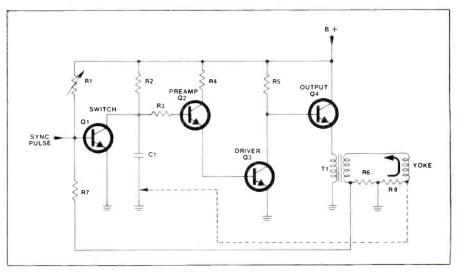


Fig. 7-A simplified Vertical Deflection Circuit.

FM Stereo Alignment

by Phillip Dahlen

Simple techniques for peaking a high-quality receiver

■ Recently a Minnesota representative of Sound Technology brought their Model 1000A FM alignment generator to Duluth to be evaluated by both the ELECTRONIC TECHNI-CIAN/DEALER lab and Mel's TV, a major dealer in town. Working in conjunction with Dennis Ford, an electronic technician at Mel's, we used this instrument to supply the desired frequency-modulated (FM) signal to the antenna terminals of a Fisher 500 receiver, observing the resulting waveforms on a Telequipment Type D54 dual-trace scope (Fig. 1).

Five test leads were required for the hookup (Fig. 2). These were

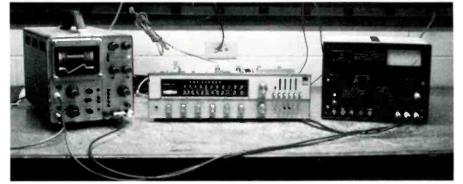


Fig. 1-Receiver and instruments used for observing FM characteristic waveforms.

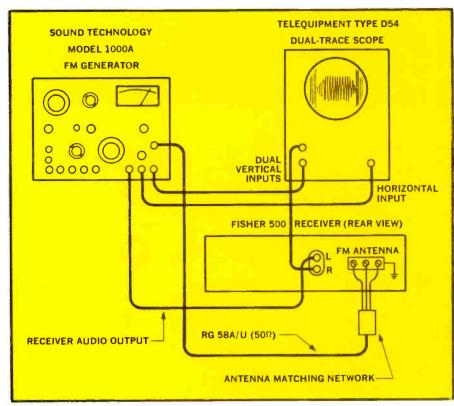


Fig. 2-Arrangement of test leads for connecting receiver and test instruments.

run between the generator's RF output and the receiver's antenna terminal, between one of the receiver's audio output channels and the generator's receiver terminal, between the generator's vertical output and one of the scope's vertical inputs, between the receiver's other audio output channel and the scope's second vertical input, and between the generator's horizontal output and the scope's horizontal input.

Unlike amplitude-modulated (AM) signals, where the audio signal is transmitted by varying the strength of the carrier frequency, FM signals are transmitted by varying the frequency of the carrier wave in response to the amplitude of the audio signal—the rate at which the carrier frequency changes corresponding to the frequency of the audio signal. The stronger the audio signal being transmitted, the greater the change in carrier frequency.

The signal generator produces both 60Hz and 10kHz signals, which are frequency modulated upon the instrument's RF signal. The oscillator control (Fig. 3) determines the amplitude of the 10kHz signal in relation to the 60Hz signal, while the sweep width control determines the amplitude of the more predominant 60Hz signal, and as a result the range of frequencies that the RF signal must cover as it is modulated. The level of the 19kHz pilot signal, used as a reference for separating the stereo channels, can also be controlled on the instrument-the test button on the instrument permitting the meter to measure the strength of this pilot signal rather than measuring the sweep width. The remaining level control determines the amount of RF signal voltage applied to the receiver, permitting a measurement



Fig. 3—Sound Technology's 1000A FM Alignment Generator.

of the receiver's effective sensitivity.

In addition to these level controls, the instrument has controls for determining the carrier frequency (some receivers are more sensitive at one end of the tuning dial than at the other), how this frequency is modulated (whether it is merely a carrier signal or contains a monaural signal, a stereo signal or an SCA 67kHz subcarrier signal, and the resulting RF waveform), and what audio signals are used (whether an external left and right channel signal is substituted for the 10kHz signal, whether only the left or right channel audio signal is transmitted, or whether a left minus right channel or right minus left channel stereo signal is transmitted).

Besides the four test-lead terminals shown connected in Fig. 2, the two left terminals on the instrument are for applying an alternate audio signal to be transmitted in place of the 10kHz signal, the third terminal can provide a 19kHz signal, the fourth terminal can provide a 10kHz signal, and the last terminal can provide a composite audio signal.

So much for virtually all of the controls and connections, but what can they be used to show?

When the generator applies a $1.7\mu v$ RF signal to the receiver (Fisher's rating for the receiver's sensitivity), the lower scope trace (obtained directly from the receiver) forms a straight diagonal line (Fig. 4)—its output voltage going

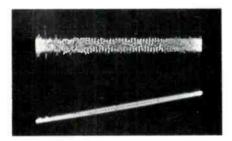


Fig. 4—Scope traces that resulted when a 1.7μ v RF signal, sweeping 150kHz, was applied to the receiver. (Upper trace is 10kHz signal obtained through generator while lower trace is 60Hz signal obtained directly from receiver.)

up as the sweep frequency rises. A small ripple may be seen on this lower scope trace. It represents the superimposed 10kHz audio signal.

In the two previous articles describing RF signals ("Tuned Circuit Signals," Parts I and II), we stressed the fact that there are occasions when it is necessary to apply a sine wave to a scope for its horizontal trace rather than using the scope's internal horizontal sweep signal (a saw-toothed wave). The scope traces shown in Fig. 4 (plus many of the other scope traces shown in this article) resulted from using a 60Hz sine wave, obtained from the generator, as the horizontal sweep signal. The phase control on the generator is adjusted so that the 60Hz vertical signal obtained from the receiver and the 60Hz scope sweep signal obtained from the generator are at such a phase angle that a diagonal line is formed. (In Fig. 22, 23 and 24 on pages 52 and 53 of the October 1970 issue we saw that when sine waves of the same frequency are applied to the horizontal and vertical inputs of a scope, a pattern forms that can look like either an oval or a circle-depending on the phase relationship between the two applied signals. The phase angles can be further shifted so that a diagonal oval, such as that shown in Fig. 24, will flatten out more and form a diagonal line-as seen in the lower trace in Fig. 4 of this article.)

As had been previously indicated, the more positive portion of the 60Hz sine wave (the higher voltage portion) is seen at the right side of the screen, while the less positive portion of this sine wave (the lower voltage portion) is seen at the left side of the screen. As the 60Hz sinewave voltage becomes greater, the modulated radio frequency becomes greater; while as the voltage forming the 60Hz signal becomes less, the modulated radio frequency becomes less.

The 60Hz signal obtained from the receiver's other audio channel is filtered out in the generator cabinet, leaving only the 10kHz signal to appear as the upper trace on the scope (Fig. 4). This 10kHz signal, which in the receiver had been riding on the more predominant 60Hz signal, is also produced by FM variations of the RF signal. Since the 10kHz signal is not nearly as strong as the 60Hz signal, its corresponding ac voltage does not produce significant frequency shifts of the RF signal (it is the amplitude of the ac signal voltage that determines the range of frequencies that the RF signal sweeps and not the frequency of the audio signal). Therefore, the left portion of the upper scope trace shows the 10kHz signal as it is produced by variations of lower-frequency radio signals, while the right portion of this trace shows the 10kHz signal as it is produced by variations of higher-frequency radio signals.

In other words, the 10kHz signal, shown as the top scope trace, is produced by FM variations of an RF signal which is changing in frequency—sweeping a larger frequency range than that required to produce this signal. The 10kHz signal shown in the left portion of the screen is produced by a lower radio frequency than the 10kHz signal shown in the right portion of the scope screen.

In this instance the amplitude of the predominant 60Hz modulating signal present in the generator resulted in a 150kHz sweep of the RF signal. This is about as wide a radio frequency range as the receiver could handle with the current RF input voltage and the present adjustment of its IF and discriminator tuned circuits. Note how a very slight trace of noise is apparent at either end of the two scope traces.

These same general principles apply to most of the remaining scope traces.

When increasing the amplitude of the generator's internal 60Hz modulating signal so that the RF signal sweeps a 300kHz frequency range, the receiver is unable to handle so large an RF fluctuation and a distorted 60Hz waveform is produced at its audio output—noise appearing at

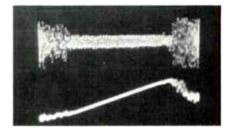


Fig. 5—Scope traces that resulted when a 3.0 μv RF signal, sweeping 300kHz, was applied to the receiver.

the left and right ends of the lower scope trace (Fig. 5) rather than an extended horizontal line representing peak maximum and minimum 60Hz voltages.

Under these same conditions, the receiver is also unable to receive the 10kHz signal carried by the higher and lower radio frequencies, and noise is therefore also apparent at either end of the upper scope trace.

In addition to increasing the RF signal sweep to cover 300kHz, we elected to increase the RF signal strength from $1.7\mu v$ to $3.0\mu v$. Although so small an increase was not adequate to help the receiver cover the entire RF sweep, it did help improve the receiver's coverage. By increasing this RF voltage to $30,000\mu v$ we were able to cover the entire

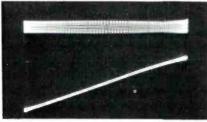


Fig. 6—Scope traces that resulted when a $30,000\mu\nu$ RF signal, sweeping 300kHz, was applied to the receiver.

300kHz RF sweep (Fig. 6) with the receiver.

By adjusting the generator for a $100\mu v$ RF signal sweeping 600kHz, the receiver distortion (Fig. 7) approximated that previously received with the lower RF signal strength and narrower sweep width. How-

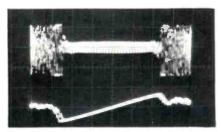


Fig. 7—Scope traces that resulted when a $100\mu\nu$ RF signal, sweeping 600kHz, was applied to the receiver.

ever, in this instance increasing the applied RF signal strength to $30,000\mu v$ did not result in the receiver adequately covering the applied FM signal (Fig. 8). (The left portion of the upper scope trace indicates that noise is present at lower

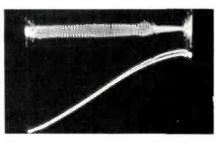


Fig. 8—Scope traces that resulted when a $30,000\mu\nu$ RF signal, sweeping 600kHz, was applied to the receiver.

radio frequencies. The right portion of the lower trace indicates that the receiver is unable to receive the more positive portion of the 60Hz sine wave and there is a corresponding loss of 10kHz signal on the right portion of the upper trace. Noise is apparent near the right end of the upper trace.)

Adjustments such as these can be used to determine a receiver's sensitivity, effective RF bandwidth and the efficiency of its automatic gain control (AGC).

This instrument can also be used to improve the receiver's IF and discriminator circuit alignment.

In Fig. 6 we saw the scope traces that resulted when the discriminator properly adjusted and a was 30.000µv signal was applied having a 300kHz sweep width. With the discriminator still properly adjusted, a similar set of scope traces appeared when 300µv of RF signal was applied having a 150kHz sweep width. By misadjusting the bottom tuning slug in the receiver's discriminator circuit, we obtained the distortion shown in Fig. 9. (For this and the following illustrations the scope's vertical sensitivity to the

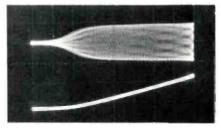


Fig. 9---Distorted scope traces that resulted when a $300_{\mu\nu}$ RF signal, sweeping 150kHz, was applied to the receiver and the lower portion of the discriminator circuit was tuned out of alignment.

10kHz signal voltage was increased so that we could more readily see how the strength of this signal was made to vary by misadjusting the discriminator circuit.) As misadjusted, the receiver became insensitive to the lower-frequency radio signals —no 10kHz signal appearing at the left end of the upper trace and the 60Hz signal's peak minimum voltage being clipped off.

With the bottom discriminator tuning slug readjusted and the top one tuned out of alignment, we ob-

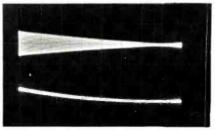


Fig. 10—Distorted scope traces that resulted when a $300_{\mu}v$ RF signal, sweeping 150kHz, was applied to the receiver and the upper portion of the discriminator circuit was tuned out of alignment.

tained the scope patterns shown in Fig. 10. In this instance the receiver was more sensitive to lower radio frequencies than higher ones and there was a stronger 10kHz signal appearing at the left end of the upper scope trace than in the central portion or at the right end.

Had the RF sweep width been even greater than 300kHz and the discriminator circuit further detuned, we would have seen a greater increase in the 10kHz signal at the right end of the upper scope trace, the minimum 10kHz signal amplitude appearing near the center of this upper scope trace. Even in Fig. 10 we can see a slight increase in 10kHz signal amplitude at the right end of the upper scope trace, as compared to one division to the left on the scope's gradient.

We did not elect to further misadjust this receiver's discriminator circuit, but had we done so we would have seen scope traces similar to those shown in Fig. 11. Such a pattern (had it occurred with the previous scope settings) would have indicated that the receiver was more sensitive to the higher and lower frequencies swept by the RF signal *continued on page 60*

Sweep Alignment Pointers

by Jim Smith

Although many articles have been written in the past on how to use a specific sweep generator or how to align a certain TV chassis, the reasons for the shape of the IF amplifier response curve and important related information have generally been lacking. It is important to know why the response curve is shaped the way it is and what effects will occur when it is not shaped properly. Understanding a few basic points about sweep alignment will make the alignment of any TV set both faster and easier

■ The IF amplifier response curve is the most important response curve in a TV set. The shaping of the signal in the IF strip will determine whether the receiver will pass color, sync in properly, have ghosts or ringing in the picture. It is also in these circuits where the greatest change will oc-

ages.

cur as the set

amplifier re-

sponse curve is

shown in Fig. 1.

Although it is

not a "text-book curve" (as in

Fig. 2), it does

come very close.

Some of the new-

er receivers, such

as Zenith have a

An ideal IF

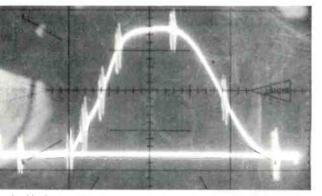


Fig. 1-Ideal IF amplifier response curve obtained from a TV set.

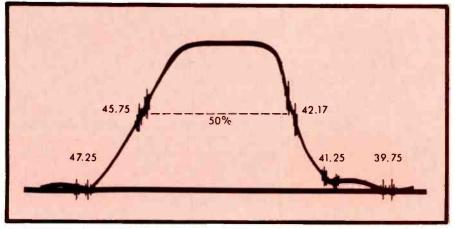


Fig. 2--- "Text-book" type illustration of what an ideal IF amplifier response curve should look like.

rounded or "hay stacked" curve.

For most receivers it is best to have the alignment instructions so that the actual shape of the curve is known as well as the location of the adjustments in the receiver. However, when doing alignment work it is best to strive to get as close to the ideal curve as possible without spending too much time on the alignment. In most cases, coming "close" can result in a fairly good job of alignment, but the more alignment experience gained, the easier it is to approach the ideal curve.

Marker Positions

The 45.75MHz video carrier signal should be seen halfway up the leading edge of the response curve. However, this marker position for that frequency may vary between 40 and 50 percent of maximum amplitude without affecting the picture. If the marker falls below the 40 percent level, some difficulty may be encountered with the sync; while if it rises above 50 percent, ringing in the picture, tunable ghosts and a poor picture will result.

If the marker appears as in Fig. 3, the slope of the video portion of the response curve will not be straight between the 47.25MHz and 45MHz points. The TV set will then experience the same difficulties as those previously mentioned,

The position of the 45MHz point on the response curve is not critical and can vary slightly just as long as the video portion of the curve remains horizontal. This point can be easily found since it will fall on the curve exactly halfway between the 45.75MHz video carrier and the 44.25MHz marker. With some of the haystacked curves in the newer receivers, it is normal for the 44.25 MHz marker to appear near the top of the curve on the video carrier slope.

There are some cases when the 47.25MHz adjacent-sound trap may be misaligned without causing any trouble in the TV set. However, Fig. 4 shows a condition in which the trap can cause poor sync conditions by reducing the straightness of the video slope on the response curve, as well as reducing the low-frequency signals applied to the video detector. The alignment of this tap is

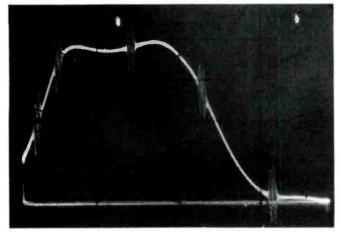


Fig. 3—A dip at the top of the IF amplifier response curve may result in ringing in the picture, tunable ghosts and a poor picture.

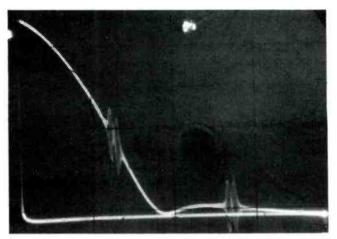


Fig. 4.—The sound trap can cause poor sync conditions by reducing the straightness of the video slope on the response curve.

very critical in TV sets connected to cable systems since they are subject to interference not normally encountered during the reception of "freeair" signals.

The color slope of the IF amplifier response curve is very important and demands as much attention in a color-TV set as does the video slope. Fig. 5 shows the positions of the color markers as they should normally appear on the response curve. If the 42.67MHz marker is allowed to fall below the 20 percent relative amplitude limit or is filtered through the 41.25MHz sound trap as a result of misadjustment, not all of the color information will reach the chroma bandpass amplifier, resulting in weak colors on the screen. However, if the 41.67MHz marker rides on the flat portion of the IF amplifier response curve, or above the 80 percent point as shown in Fig. 6, the same trouble may result but with the additional possibility of ringing or a ghost effect in the picture. Slight variations in the color slope portion of the response curve can be compensated for with the chroma bandpass amplifier if the markers are close to their proper location.

The chroma bandpass amplifier has an unequal response that is opposite that of the color slope of the IF amplifier. As a result, the response curve of signals at the output of the chroma bandpass amplifier is flat, and information of uniform amplitude is fed to the color demodulators.

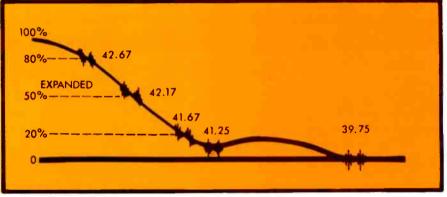


Fig. 5---Relative positions of the color markers as they should normally appear on the response curve.

Alignment Procedures

The traps in a TV set are very important and should be adjusted before any other alignment is performed. Generally, the 41.25MHz sound trap is the one that will be found to need a slight touch up. Misalignment of this trap will cause sound bars to appear when the picture is properly tuned in, or a smeary picture if the sound is tuned out of the picture. The adjacentsound trap will upset the slope of the video curve and, as explained previously, also cause smear in the picture.

Alignment of the 41.25MHz sound trap can be made easier by tuning to center the 41.25 marker and setting the scope's sweep width to about 1.5MHz. In this manner the IF amplifier response curve is enlarged and its dip becomes more apparent for trap adjustment. From the resulting scope trace, shown in Fig. 7, we see the 41.25MHz marker above the dip caused by the trap. This indicates that the trap is out of alignment and that in this case the sound signal is not being filtered out as it should.

Fig. 8 shows a chroma bandpass amplifier response curve with the 3.08MHz and 4.08MHz markers at equal amplitude. This is the response curve seen when the sweepmarker signal is injected into the IF test point in the tuner and the signal is passed through both the IF and chroma bandpass amplifiers. If instead the marker-sweep signal is injected into the chroma bandpass amplifiers, as recommended by several manufacturers in their older alignment procedures, the unequal response curve shown in Fig. 9 will result. This unequal response of the

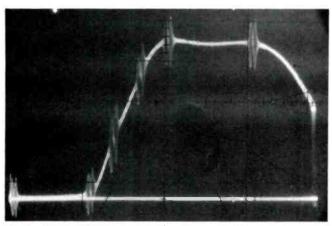


Fig. 6—If the 41.67MHz marker rides above the 80 percent point not all of the color information will reach the bandpass amplifier and there may also be ringing or a ghost effect in the picture.

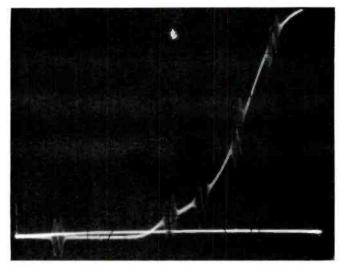


Fig. 7—The 41.25MHz marker appears above the dip caused by the trap, which indicates that the trap is out of alignment and that in this case the sound signal is not being filtered out as it should.

chroma bandpass amplifier matches the unequal response of the color slope on the IF amplifier response curve, as explained earlier. The best procedure, which is also recommended by most manufacturers today, is to align the chroma bandpass amplifiers for a flat output curve when passing the sweep-marker signal through both the IF amplifier and the chroma bandpass amplifier just as the normal color signal would have to pass through both. This results in greater accuracy when aligning the chroma bandpass amplifier.

There has been some question concerning the 4.5MHz sound trap since many procedures are given for aligning this trap which fail to include any explanation as to what is happening. The 4.5MHz trap is located after the video detector and cannot be aligned with the normal IF sweep-marker curve. However, this trap can be aligned by watching the chroma bandpass amplifier curve and the 4.5MHz marker; and then adjusting for a minimum gain from the base line of the marker or by watching the dip as was done when aligning the traps in the IF response curve. This makes it a simple trap to align after learning where it is and when to align it.

Conclusion

Sweep alignment is a very important tool as it can correct the picture in many TV sets that have troubles which seem "sort of way out." But sweep alignment, like anything else,

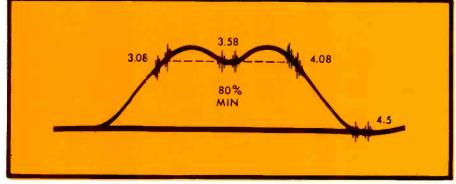


Fig. 8—By feeding the sweep-marker signal through both the IF and chroma bandpass amplifiers, the 3.08MHz, 3.58MHz and 4.08MHz markers appear to all be of equal amplitude.

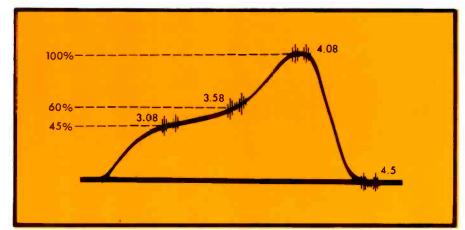


Fig. 9—By feeding the sweep-marker signal through only the chroma bandpass amplifier, an unequal response curve results. This curve matches the unequal response of the color slope on the IF amplifier response curve.

must be practiced if it is to be done proficiently. With more practice, it can be done faster, and the customer will be pleased with a better job. Providing customers with cleaner and sharper pictures will more than pay for itself in days to come.

GUEST AUTHOR

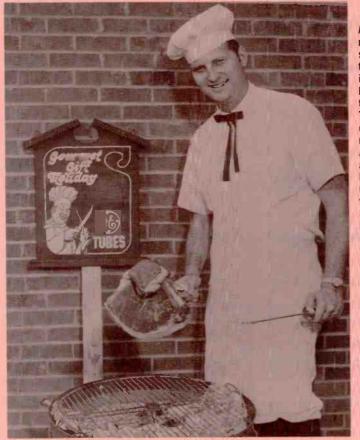
The Independent Dealer-Distributor Relationship

by Arthur M. Effron

Keeping this relationship satisfying and viable is not only a challenge to the manufacturer's marketing effort but also a daily test of every supplier's professional concern for future market growth.

■ Although our approach to market promotion emphasizes a close relationship between the independent service dealers and the distributor, compatibility here is not entirely contingent on product line, pricing and service. There is another element that is certainly worthy of attention-incentives.

Sales incentives are used for many reasons. Among these is the need to help remind dealers and distributors that they are receiving advertising and service support, plus high-quality, reliable products that are supplied for their customers.



Art Effron makes ready a prime steak from one of the gift offerings in General Electric's promotional program. The author is a district sales manager in Los Angeles for General Electric's **Distributor Sales** Operation, Electrenic Components Sales Dept. Prior to this appointment, he was manager of advertising and sales promotion for General Electric's Tube Products Dept. in Owensboro, Ky. He is a graduate of the University of Cincinnati and has sment more than nine years in the electronics industry.

The best incentives are not found in novelty shops, but instead they should feature the same built-in quality as the product line sold. Incentive programs are usually keyed to major holidays, vacation periods and sports for greatest appeal. Such programs work particularly well during prime selling seasons.

The key to the success of many promotions today is coming up with quality, name-brand "bonus items" or incentives with *immediate* attraction. Above all, do not offer junk.

Items that we have recently used successfully include luggage; namebrand cameras, pocket radios and hams; and coupons redeemable for gas at national-brand stations. Each had instant appeal and was convenient and easy to use.

Now I suppose it would be easy to stop here and ignore such "minor" points as what the distributor really wants or needs in terms of his local market, or whether or not he will consider one line of products over the competitors. After all, the market effort is one thing, while your product image is another.

As we well know, distributors often follow the path of least resistance and "push" a product simply because it is overstocked, there is a promotion program going on, they have just been briefed on how to sell it, or it is "new." Or, they will sell the dealer the brand he specifies, without even thinking to recommend a better value.

We think this viewpoint can be changed by combining quality services and products with quality incentives—often something of value, something worth working for, something worth taking home to the family. Better yet, offer more than one item.

There is no doubt in my mind that dealers and distributors often are quite selective in over-thecounter marketing. But once the dealer and distributor taste the fruits of a sound, well-planned marketing program offering the best in incentive products and sales opportunities, they will certainly consider the corresponding manufacturer's products again when placing future orders. And that is what it is all about.

TEST INSTRUMENT REPORT

Heathkit Model IB-101 Frequency Counter

by Phillip Dahlen

Operational simplicity was apparently one of the goals of the instrument's designers

Obtaining precise frequencies is a must for today's electronic technicians. FM multiplex receivers must produce 38kHz signals that are in phase with the pilot signal, color TV sets must generate a 3.58MHz continuous wave signal in phase with the color burst signal. And last month's article, "Commercial Two-Way Radio," told of all the frequencies available for business use, which must be precisely tuned. These and many other similar applications require the availability of a precise frequency counter for calibration purposes.

The IB-101 counter uses digital

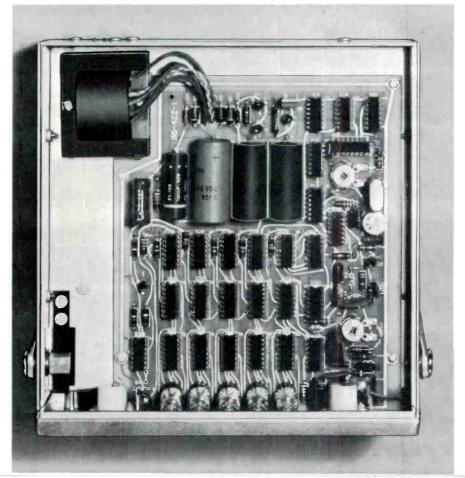


Fig. 1—Heathkit's Model IB-101 frequency counter. For more details circle 900 on Reader's Service Card.

integrated circuitry to provide counting from 1Hz to reportedly over 15MHz with the elimination of divider chain adjustments. Specifications indicate that an overrange indicator and Hz/kHz switch gives the instrument's five cold-cathode display tubes the same capability as a more expensive eight-digit counter. This means that with the use of that range switch you should be able to make such frequency readings as 14,857,236Hz (a number taken at random to indicate the possible capability). The input circuit uses a dual-gate, diode-protected MOSFET that is designed to provide proper triggering levels from less than 100mv to greater than 200v without any input level adjustment. The input impedance is said to be 1M shunted by less than 20pf to minimize circuit loading and error.

The \$199.95 kit includes a double-sided circuit board, dual primary power transformer, 26 integrated circuits, 7 transistors and 5 cold-cathode display tubes, plus many related components.

Fig. 2—Although the instrument contains complex circuitry, components are well arranged to simplify construction.



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"We advertise under washer-dryers, television, and vacuum cleaners. That way, whatever a potential customer needs, he's bound to run across us" explains Albert Carlson, of Barn Appliances, Fenton, Michigan. "In a small area like this you have to build up a good reputation, and you have to stay in contact with people. One of the best ways to do that is in the Yellow Pages. We stress service after a sale. It's one of our key points in our Yellow Pages ad. That way customers know we stand behind what we



sell. We've been advertising in the Yellow Pages for 13 years now, and its brought us nothing but good customers." Let the Yellow Pages do your talking. People will listen.



An effective way to build business.

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N	lodel	Max, Payload	Max.GVW
E	-300 -200 -100	4320 lbs. 1800 lbs. 1120 lbs.	8300 lbs. 5400 lbs. 4500 lbs.



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maneuverability in city delivery operations --time saved on every trip. So

Wider at top for built-ins.

Body sides are more vertical, wider apart at top than other vans. So built-in units fit better and leave more aisle. Modular units, designed to fit and work together allow you to custom

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FORD ECONOLINE VANS

A better idea for safety: Buckle up

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TECHNICAL DIGEST

The material used in this section is selected from information supplied through the cooperation of the respective manufacturers or their agencies.

OLYMPIC

Table Radio Models 32/33/34/35-Service Hints

Symptom: Weak sound output, no distortion discernible. Correction: Check the following electrolytics by substitution: C47, C55 and C56.

Symptom: FM inoperative.

Correction: Check transistor Tr-1 and the following electrolytics by substitution: C3 and C10.

Symptom: AM reception weak, FM inoperative.

Correction: Check the following electrolytics by substitution: C27, C33 and C19. Also consider a transistor defect whenever the forward bias exceeds 0.4v.

Solid-State TV Model 9P59—Service Hints

Symptom: Loss of raster within first 30 min. of operation. Correction: Replace resistor R514, a 1.5Ω , $\frac{1}{2}w$ base bias resistor in the horizontal output stage.

Symptom: Critical vertical sync lock and/or complete loss of vertical and horizontal sync.

Correction: Check sync transistors Tr-18 and 19. Also check by substitution capacitor C405, $10\mu f$ at 25v.

Symptom: Picture flagging at top of raster.

Correction: Replace capacitor C503, 3µf at 6v located in

OSCILLOSCOPE/VECTORSCOPE

MODEL TO-50

- DC to 10 mhz frequency response
- .02 volt sensitivity
- Calibrated vertical attenuator
- Calibrated time base
- Supplied with combination direct/locap probe
- 5 X magnifier
- Automatic triggering mode
- 5" flat face tube edge-lit graticule



... for more details circle 118 on Reader Service Card

Anti-Hunt network of the Horizontal AFC circuit.

Symptom: Picture bending and distorted.

Correction: Replace capacitor C805, a dual section electrolytic in the power supply.

Symptom: Weak sound output.

Correction: Check the following electrolytics capacitors by substitution: C323 and C320 located in the audio-amplifier circuitry.

Symptom: Fine grain appearance in video (4.5MHz beat). Correction: Realign the L201-trap coil. Refer to service manual for alignment instructions.

Model 327 Stereo Chassis-Service Hint

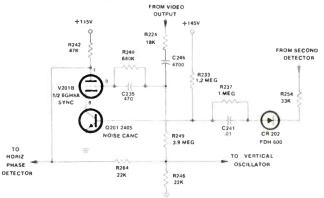
Symptom: Whenever this chassis is separated from cabinet wiring harness, as during shop service, it will be necessary to install a jumper wire between pins 2 and 4 of the remote speaker socket to secure stereo speaker operation.

RCA SALES CORP.

TV Chassis KCS179/KCS183—Sync and Noise Cancellation Circuit

This receiver's sync section consists of the sync-separator stage and the noise-cancellation stage. Under normal signal conditions, the sync-separator tube, V201B, and the noise-cancellation transistor, Q201, are conducting.

A positive-going signal from the video output stage is coupled through resistor R224 and capacitor C246 to the RC network made up of resistor R248 and capacitor C235.



The signal bias developed by this RC network sets the operating point of the tube, allowing the tube to conduct during sync-pulse time only.

A negative-going video signal from the second detector is coupled through resistor R254 and diode CR202 to the RC network, consisting of resistor 237 and capacitor C241. This RC network sets the cutoff point of the transistor at the sync-tip level.

Under normal signal conditions, both stages are conducting, providing sync pulses to the vertical and horizontal oscillators. Since transistor Q201 is operating just short of cutoff at the sync top level, any pulse, such as a noise spike, which drives the base of the transistor more negative then the sync tip will cut off the transistor. With transistor Q201 cut off during a noise spike, the conduction path of the sync amplifier tube V201B is open, deleting any output from the sync separator during the noise-pulse time; thus maintaining stable vertical and horizontal synchronization.

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reference and guide for electronic technicians who need to understand and repair semiconductor circuits efficiently. For those interested in transistor physics, fundamentals are emphasized in the first two chapters. The real 'meat'' begins in Chapter 3 which will thoroughly familiarize you with amplifier fundamentals, basic circuit configurations, biasing, FETS, JFETS, and IGFETS. The next two chapters will acquaint you with RF and IF amplifiers. 256 pps., over 150 illus., 12 Chapters. Order No. 504 List Price \$7.95 .

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the 1-watt phono amp and IC power supplycan be built in an evening. More sophisticated projects—like the electronic organ or the RDIAA equalization preamp-offer a greater challenge. You can build practical devices like the tachometer with bulb alert, or the 50-watt amplifier, or some "just for fun" gadgets like the simple memory tester or the miniature adding machine. 160 pages, 50 projects, 100 illus.

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FLECTRONIC HOBEVIST'S

PROJECTS HANDBOOK

Electronics Reference Databook

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This new book is much more than a simple collection of tables, formulas, graphs, equations. etc. In addition to the abundance of helpful information given, it provides specific guidance in the use of data. Numerous problems associated with every level of interest-from electronics theory (formulas, laws) to measurements, tests, and circuit

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design work-are covered. In so doing, the author explains how to use the data (from this or other volumes) for purposes other than those listed. Covers Electronics Theory, Use of J Operator, Exponential and Other Tables. Attenuators and Equalizers, Filter Design, Practical Component Design and Application, Tube and Semiconductor Circuit Design and Operation, and Transmission Lines. 224 pps., over 100 illus., plus 45 tables. Hardbound.

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How to Repair Solid-State Imports



Among this vast lection of nearly 100 hard-to-find schematics you'll find such names Allied, Automatic as Radio, Peerless Tele-rad, Mercury, Lloyd's, Panasonic, Crown, Midland, Penncrest, Toshiba. and Matsushita. representing the major chassis foreign-made distributed in the U.S. And to help you find replacement parts, there

is a list of importers and distributors of Japanese consumer products. There's also a chapter to help you when a schematic isn't available-numerous tips to help you get by without the exact diagram. Truly unique among manuals of this type. Chapter 5 contains a host of schematics and service data represent-ing virtually every import you'll come across. along with a 24-page foldout section. 160 pps., $8\frac{1}{2}x11^{"}$, plus 24-page foldout section.

List Price \$7.95



Philco Color TV Service Manual

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An all-in-one service guide for Philco color sets, with 12 complete schematic diagrams for chassis 15M90/91 to 20QT88. Here in one manual is complete service data for all the color models produced by Philco and Philco Ford (thru 1970), from the all-tube to the latest hybrid solid-state chassis, including the small-screen portable

Model T5062WA. The unique 36-page foldout section contains 12 complete schematic dia-grams, representing all the chassis covered. The profusely illustrated text delves into each section (video, chroma, vertical, horizontal, etc.), and points out specific problems based on the author's extensive experience. Included are complete alignment and setup instructions, detailed in step-by-step form. 160 pps., plus 36-page schematic foldout section. Long-life vinyl cover. Order No. 522 List Price \$7.95 .

How to Use Test Instruments in **Electronics Servicing**



handbook on test equipment applications ranging from the use of audio gear to tube and transistor checkers. Just what you need to put your test equipment to work. Not a "how-it-works" treatment, but a "how-to" manual describing specific tests and troubleshooting techniques for the electronic techni-

A long-needed, practical

cian. You'll discover new ways to use your scope and several new "tricks" you can perform with multitesters. You'll learn signal-injection troubleshooting, how to measure inductance and capacitance with the help of your signal generator, pointers and pitfalls for using markers, sweeps and pattern generators, shortcuts and special techniques for color TV troubleshooting, how to test audio circuits and FM stereo equipment, and much more. 256 pps., over 200 illus. Hardbound. Order No. 485 List Price \$7.95 .

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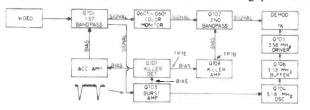
COLORFAX

The material used in this section is selected from information supplied through the cooperation of the respective manufacturers or their agencies.

ADMIRAL

Color TV Chassis K20-Troubleshooting For No Color

As shown in the block diagram, the chroma section of the K20 series color-TV chassis has 10 transistor stages. Of these, six circuits function as ac amplifiers and four circuits function as dc control switches. The following procedure



will help you to quickly isolate a defective stage for "no color" after which normal troubleshooting will isolate the defective part.

1. Apply color bar signal to VHF tuner terminals (a color program can be used).

- 2. Set Color control to mid-range.
- 3. Set Color Threshold control fully clockwise.

Test Procedure

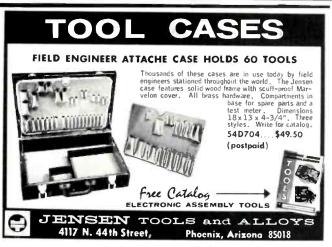
1. Connect scope to point 7K (Pin 2 of 6BV11 tube) and check for a 3.58MHz signal. (Test point 7K is the pin to the left of the word "7K" printed on the board—do not use the pin directly above the printed 7K.)

- a. If a 3.58MHz signal is not present, the problem is in the buffer, driver or color oscillator circuit. Proceed to step 2.
- b. If a 3.58MHz signal is present, proceed to step 3.

2. Using an AM generator, inject approximately a 3.58MHz signal at the base element of transistor Q706. Leave the scope at point 7K and watch for the 3.58MHz signal. Use the same procedure in checking transistors Q704 and Q705. After you have isolated the stage that does not produce a signal at point 7K, use a VTVM to locate the faulty component. Do not overlook the crystal circuit.

3. Connect the jumper from the 25v supply (red wire) to TP7B (Collector of transistor Q708, the killer amplifier) on the circuit board.

a. If the color does not appear on the CRT, the problem is in the Color Monitor (if used) or in the first or second bandpass amplifier. Continue to step 4.



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b. If color does appear, the problem is in the Killer amplifier, Killer Detector or Burst amplifier. Proceed to step 5.

4. To check the Color Monitor, disconnect the wire from point 6F on the monitor board and connect it to point 6H.

a. If color is now present, the problem is in the Color Monitor. Check the stages containing transistors Q601 and Q605.

b. If color does not appear, the problem is in the first or second bandpass amplifier. Use the VTVM to check.

5. Disconnect the jumper to point TP7B that was added in step 3.

6. To check the Killer amplifier, connect a 68K resistor from TP7E to ground (collector of transistor Q707). If color now appears, the Killer amplifier is okay. Remove the resistor.

7. To check the Killer detector, short the base element to ground. The collector voltage should rise to approximately 22v. Remove the base short and the collector voltage should decrease to approximately 17.5v. If the stage voltages are okay, proceed to step 8.

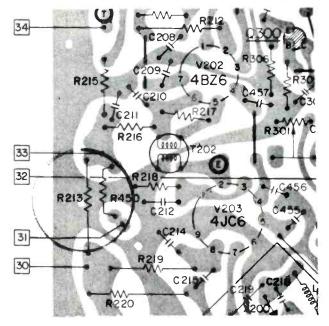
8. If the Burst amplifier is at fault, there should be color on the CRT during step 3 but it is probably out of sync.

WESTINGHOUSE

Color TV Chassis V2655, V2656-Service Hints

In early production of the IF printed circuit boards, the value of resistor R213, a 47Ω resistor, was incorrectly marked 47K on the top of the PC board.

Resistor R213 is a stone-type fusible resistor which may



break apart under overload conditions. When this happens, the technician may see the 47K marking and replace the burned out one with a 47K value (which is wrong), the correct value should be 47 Ω . Please note the location of resistor R213 on the partial diagram of the IF PC board and replace it only with a 47 Ω , 4w, fusible, stone-type resistor, Westinghouse part number 250V020H79.

Thirteen cures for 78 headaches.

Integrated circuits are replacing transistor circuits in many stereos, radios, B & W and color TVs.

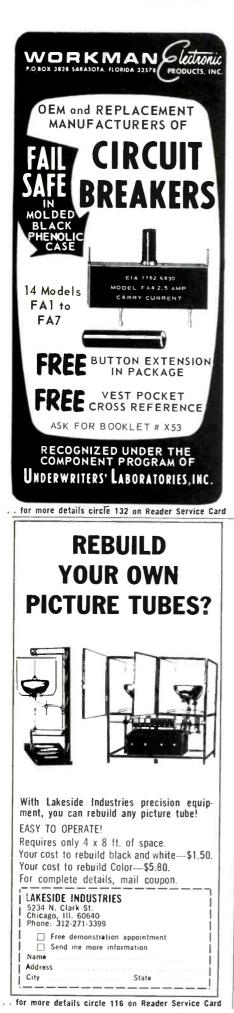
That means more parts to stock and more money tied up in inventory.

Now, Sylvania ECG can take some of the pain out of these stocking headaches.

Our 13 ICs will replace 78 part numbers, including RF, IF and audio amplifiers, sound detectors, oscillator/mixers, chroma demodulators, and automatic fine-tuning systems.

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GIÐ SYLVANIA



NEW PRODUCTS

For additional information on products described in this section, circle the numbers on Reader Service Card. Requests will be handled promptly.

703

ATTENUATOR

Replaces standard carbon components

Distortion free ac signals, 0 to 300 MHz, for CATV or audio distribution systems are reportedly now provided by a 75 Ω plug-in "T" pad, measuring 0.375 in. in diameter and 0.473 in. high, and incorporating a thick film cermet resistive element. The design reportedly provides minimum inductance with improved stability. They are said to be available in 11 values ranging from 0 to 15dB. Signal levels are determined at each CATV/CCTV or audio installation by simply plugging in the proper "T" pad. Rated tolerances are ± 0.05 to ± 1.0 dB, depending on the component. The plas-



tic-cased plug-in attenuator reportedly mounts in a standard TO-5 base having a 200 in. pin circle. The manufacturer indicates that it has gold-plated half-hard corrosion-resistant brass pins with rounded tips for easier insertion. The "T" pad is said to be installed in PC boards to replace standard carbon composition components, especially in automatic control circuits where attenuation is needed over a 0 to 300MHz frequency range. Examples of its use at audio frequencies are said to include line amplifiers, telephone terminals, communications receivers, and similar 75Ω impedance/power matched installations. Other impedances are said to be available on special order. Aerovox.

TV STRIP AMPLIFIER 704

Occupies minimum of rack space

A series of solid-state, high-voltage strip amplifiers is developed for the distribution of TV and FM signals in very large MATV and CATV systems. Manufacturers specifications indicate the unit features a full 4v output and occupies only 134 in. of rack space.



Designated the THPM, each singlechannel unit has a rated bandwidth of 6MHz for semi-adjacent TV channel operation. Models are said to be available for low band operation (Channels 2 to 6), high-band operation (Channels 7 to 13) and FM band operation with a 20MHz bandwidth. Each amplifier reportedly has separate 75Ω input terminals for various source pick-ups, while outputs are loop-mixed with 75 Ω cable. Specifications indicate that the total gain of the AGC controlled TV amplifier is greater than 55dB, while the gain of the FM amplifier is above 45dB. The wall or rack mounted units have a rated output capability of 4v (+72dBmv), with a level control to adjust between 1v and 4v. For good skirt selectivity, greater than 30dB at the next non-adjacent channel edge, the amplifiers reportedly have triple-tuned input and double-tuned output band-pass filters. Jerrold.

CONTACT CLEANER

705

Force removes solids and leaves no residue

A contact cleaner is said to contain trichloro-trifluoroethane with a fluorinated propellant

in an aerosol container. It reportedly forces residual contamination from the surface and dissolved oils are said to be volatilized and removed in seconds with solvent evaporation. Manufacturer specifications indicate that the cleaner has no flash or fine point and is nonexplosive, nonstaining and noncorrosive for all metals. The



solvent is formulated to permit clean-

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They're "market-engineered" to dominate color reception lead-in sales. Metro-Color[™] for congested viewing areas. Maxi-Color[™] for the suburban viewing market. Real sales grabbers. Created to bring home a better color picture more profit for you. Colorful eyestopping packages. Long warranties. Flexible, small size for easy installation. Competitive pricing. And a "Hot Line" display, too! That's BELDEN the program. Grab it while it's hot. See your Belden DISTRIBUTOR today. 8-3-C

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NEW PRODUCTS

ing of components without having an effect on plastics, elastomers, painted or varnished base material. Low surface tension reportedly permits penetration of the contact cleaner into cracks and crevices and also allows for good wetting properties. CRC Chemicals.

CRT ADAPTER

706

Converts tube tester into CRT tester

The Model MH-3A CRT Adapter reportedly tests all B/W picture tubes ranging from 8 in. to 30 in., 50° to 115°, 12 pin bases, 8 pin bases or the



very latest 7 pin bases. Specifications indicate that two 14-pin sockets and a special color switch are designed to check each of the red, green and blue color guns of the color CRT separately. Coletronics Service.

νом

707

Large three-color meter panel with mirror-scale

The WV-520A VOM features a rated sensitivity of $100,000\Omega/v$ for dc measurements and $10,000\Omega/v$ for ac measurements. Also featured is a



rotating function and range switch, a convenient polarity reversing switch, and a large meter panel with a mirrorscale. Manufacturer specifications indicate that this instrument is designed to

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will carry a 12-foot sailboat , plus its mast
a living room-size roll of carpeting , a living room-size roll of carpeting , a snooker table , or a full-size alligator
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it the world's biggest compact van.
Because that's what the Dodge Maxivan is.
Open its doors and say, "ah!" **NEW DODGE**



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NEW PRODUCTS

measure dc voltages from 10mv to 1,500v, ac voltage (RMS) from 100mv to 1500v, direct current from 0.5ma to 5amp, resistance from 0.05Ω to 20M and decibels from -20dB to +55.5dB. The instrument is said to be in a case that measures 7½ in. by 5¼ in. by 2¾ in., weigh 3 lb and is reportedly supplied complete with test leads and two 1.5v C-cell batteries. Price \$48. RCA.

SIGNAL-TONE ENCODERS 708

Five switch-selected output tones

Automatically activated signal-tone

FM STEREO...

continued from page 40 than it was to those RF frequencies that fell between. Under these conditions, a receiver would be able to tune in the same station at two adjacent points on the dial.

As had been previously indicated, the description just given for the scope traces shown in Fig. 11 would have been true had the scope been functioning as before. Actually something entirely different was seen

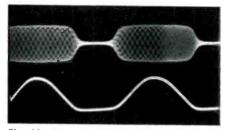


Fig. 11—Same vertical signals applied as in Fig. 9, but with the scope's 30Hz internal horizontal sweep rather than the 60Hz sine-wave horizontal sweep provided by the generator.

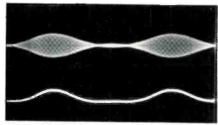


Fig. 12—Same vertical signals applied as in Fig. 10, but with the scope's 30Hz internal horizontal sweep rather than the 60Hz sinewave horizontal sweep provided by the generator.

on the scope. The scope traces in Fig. 11 and 12 are of the same signal conditions illustrated in Fig. 9 and 10, respectively, except that the

encoders are designed for use with two-way radio equipment. Keyed by closing the transmitter microphone switch, the encoder generates a short duration tone burst which modulates the transmitter and automatically ac-



scope was using an internal 30Hz saw-tooth horizontal-sweep signal rather than the 60Hz sine-wave horizontal-sweep signal that had been used. In the previous illustrations we saw a distorted 60Hz sine wave folded back upon itself, and with the non-linear horizontal scope trace it appeared as a distorted diagonal line. (This was not due to some fault of the scope. Sine waves are non-linear in nature.) These last two illustrations instead show a pair of waves representing the distorted 60Hz signal obtained from the detuned discriminator circuit. A similar comparison could be made of the 10kHz patterns-the top scope traces in each of these illustrations.

When aligning an FM receiver, the IF and discriminator circuits should be adjusted so that the diagonal wave shown in the bottom scope trace (Fig. 4) remains straight, while there is little noise apparent in the upper scope trace. With each improvement in alignment, it is possible to reduce the generator's RF output while maintaining the desired scope pattern thus increasing receiver sensitivity.

Many multiplex waveforms are also possible.

Space does not permit us to cover each of the many special features incorporated in this generator. Because of such limitations, the remaining photographs show but two of the many stereo waveforms that

tivates tone-access repeaters, tone operated receiver squelch circuits, or other signaling devices. The Model TE-5 is said to have five switch-selected output tones. The frequency range is reportedly field adjustable from 1600 to 2800Hz with a frequency stability of ± 2 percent from -10° to +60°C. Specifications indicate that the 0.5 sec output tone burst is a 6v P-P open circuit sine wave, with output impedance adjustable in three steps. The encoders and their 9v batteries are reportedly housed in a 21/8by 11/8- by 31/4-in. high-impact plastic case and connected to the transmitter microphone audio circuit by a 15in. shielded cable. An aluminum bracket is provided for mounting the encoder to the transmitter. Price \$39.95. Ross and White Co.

this generator can produce. Fig. 13 shows the right channel receiving a 60Hz signal (note the 10kHz signal riding upon it), while Fig. 14 shows

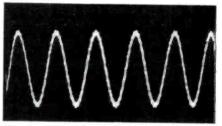


Fig. 13—Waveform observed when a 60Hz stereo signal was transmitted to the receiver's right channel, the channel to which the scope was connected.



Fig. 14—Virtually no waveform was observed when the same 60Hz stereo signal was transmitted to the receiver's left channel, while the scope remained connected to the right channel.

the same audio channel after the stereo signal has been switched to the left channel. In this instance the channel separation resulted from a 10 percent pilot signal level, but channel separation was also observed with a 2.5 percent pilot signal level.

As with the previously described alignment procedures, stereo separation can be improved by peaking the tuned circuits to maintain the desired waveform while at the same time reducing the generator's signal output.



Model 310-FET

If you need a high-sensitivity, hand-size V-O-M... Buy Triplett's 310-FET

- Hand-size FET V-O-M with 10 megohm DC input resistance.
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accuracy, a polarity-reversing switch and a rugged suspensiontype meter to soak up the hard knocks make the 310-FET the most convenient and most capable hand-size V-O-M you can buy.

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With a 10 megohm input resistance and a 300 mV DC sensitivity, Triplett's handy little Model 310-FET can handle practically any electrical measurement you may need.

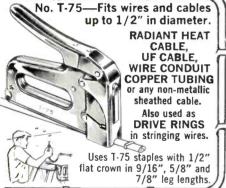
Seventeen ranges (plus 6 AC current ranges to 300 A with its optional clamp-on ammeter attachment), 3% DC and 4% AC



Blade stops staple at right depth of penetration to prevent cutting into wire or cable insulation! No. T-18-Fits wires up to 3/16" in diameter. BELL TELEPHONE, THERMOSTAT, INTERCOM, BURGLAR ALARM and other low voltage wiring. Uses T-18 staples with 3/16" round crown in 3/8" leg length only. No. T-25—Fits wires up to 1/4" in diameter. Same basic construction and fastens same wires as No. T-18.

Also used for RADIANT HEAT WIRE

Uses T-25 staples with 1/4" round crown in 9/32", 3/8", 7/16" and 9/16" leg lengths. T-18 and T-25 staples also available in Monel and with beige, brown and ivory finish at extra cost.



Arrow Automatic Staple Guns save 70% in time and effort on every type of wire or cable fastening job. Arrow staples are specially designed with divergent-pointed legs for easier driving and rosin-coated for greater holding power! All-steel construction and high-carbon hardened steel working parts are your assurance of maximum long-life service and trouble-free performance.

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DEALER SHOWCASE

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709

TOOLS

Display board with walnut finish and colorful header

A compact, specialized tool department, No. 70071, is designed for high performance. Reportedly included on the display are reversible screw drivers, screw launchers, spring tools, selfadjusting nut drivers, 6/12v testers, crimping tools, flexible shaft tune-up tools, hex key wrench sets, pow'riveter, wire strippers, and a selection of solderless terminals and pow'rivets. The display panel features are said to include: on-sight inventory control through printed illustrations and stock numbers of each item, attractive wal-



nut finish with colorful header and durable chrome-plated hanging hooks. The display is said to measure 24 in. by 36 in. and sell for \$100. Vaco.

CAR STEREO TAPE PLAYER

Total rated peak power of 18w

The Model 12R300 eight-track tape stereo player features solid-state circuitry and operates on 12vdc negative

710



gnd. It is rated for a total peak power of 18w and has a front loading arrangement with dust door. Controls include push-button program selector and sliding Stereo Balance control with center detent, as well as safety recessed volume and tone thumbwheels. Price \$78.50. RCA.

DIRECTION FINDER RADIO 711

Covers 30 short-wave services plus marine and aviation

The CR-44A "Ranger" is designed to tune over the 200 to 400kHz range where directional finding (DF) beacon stations transmit. The finder is reportedly equipped with an AGC circuit and special 20dB transistor amplifier used in the meter circuit to provide null indication. The portable's high efficiency circuitry reportedly provides for extended battery life (400 hr), precise signal-locking reliability and over 500mw of audio output power. The radio is also said to provide AM reception of standard broadcast stations on 540 to 1600kHz; FM in the 88 to 108MHz band; FAA and Canadian weather-navigation stations, plus



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(B) \$34.95* Buys A Portable Volt-Ohm Milliammeter. Measures AC & DC volts 1.5-5000 full scale. DC current from 150 uA to 15A. Resistance midscale from 15-150,000 ohms. Large 4½″ 50 uA movement meter for extra accuracy. MM-1, 5 lbs.

© \$29.95* Buys An Accurate VTVM. 7 AC & DC ranges measure RMS volts from 1.5-15,000 full scale ... AC P-P from 4.0-4000 ... 7 resistance ranges from 0.1 ohms to 1000 megohms. 25 Hz – 1 MHz response. Single probe makes all measurements. IM-18, 5 lbs. Assembled IMW-18, 6 lbs. ... \$54.95*

\$41.95* Buys A Laboratory AC VTVM. Especially useful for low-level AC & audio work. Ten RMS ranges from 0.01–300 V full scale... measures dB from -52 to +58. ±1 dB response from 10 Hz-500 kHz. 10 megs. input impedance. IM-38, 5 lbs. Assembled IMW-38, 6 lbs... \$57.95*

(E) \$39.95* Buys A Big Service Bench VTVM. Has the same high performance as the IM-18 above, plus added features to make it more useful for service work... separate 1.5 & 5 VAC scales ... calibration controls that are adjustable from the front panel ... versatile gimbal mounting ... large 7" meter. IM-28, 7 lbs.

(€) \$49.95* Buys A Big Solid-State Volt-Ohm Meter. Battery-powered portability plus built-in AC supply. 8 AC & DC ranges 0.5-1500 full scale ... 7 resistance ranges (10 ohm center scale) x1-x1 meg. High input impedance & 6" meter for greater accuracy. IM-16, 10 lbs. Assembled IMW-16, 11 lbs....\$79.95*

© \$85.00* Buys A Deluxe Solid-State Volt-Ohm Milliammeter. 9 AC & DC ranges from 150 mV-1500 V full scale ...7 resistance ranges measure from 1 ohm to 1000 megohms...11 current ranges from 15 uA-1.5 A full scale. 100 kHz response...high input impedance... large 6" meter with zero center. IM-25, 11 lbs. Assembled IMW-25, 11 lbs...\$120.00*



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THE WORKHORSE

The industry's top replacement tube, 6GH8A, is used in so many makes of TV and in so many different applications that it will be a high volume replacement type for years to come.

RCA's versatile 6GH8A is designed to satisfy the demands of all these applications:

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DEALER SHOWCASE

marine and aviation radio beacons on the 185 to 400kHz long-wave range.



The radio reportedly also provides the user with marine calling and distress, marine weather stations, amateur and other services in the 2.0 to 5.2MHz range. Services in the 152 to 174MHz range are also said to be received. The manufacturer indicates that this is a 7½-lb, "D"-battery-operated radio which is equipped with a navigational direction finder that operates on three bands: longwave, standard broadcast (AM) and marine 2.0 to 5.2MHz. Retail price \$149.95. Hallicrafter.

712

ANTENNA MOUNTING Hardware

Blister packed for merchandising convenience

A line of TV-antenna mounting accessories including masting, ground rods and UHF/VHF transmission lines is blister packed for merchandising convenience. The line is said to include chimney mounts, roof mounts, wall mounts, standard lead-in wire insulators, turnbuckles, guy hooks, eye bolts, guy rings and clamps, lightning



arrestors, ground rods, various lengths of poly clad transmission line, signal splitters, multiset couplers, and lengths of mast. Channel Master.

713

BURGLAR ALARM

Monitoring capability expands for broader coverage

A solid-state ultrasonic burglar alarm system contains a separate receiver and transmitter for greater flexibility and broader coverage. Using the same detection principles found in military and industrial alarm systems,



the patented "double check" electronic circuitry reportedly offers built-in protection against false alarms. The unit is designed for usage in small commercial businesses as well as homes and apartments. Northern Electric. IF YOU ARE A TY REPAIR MAN YOU NEED **V TECH AID**

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TECHNICAL LITERATURE

For additional information on products described in this section, circle the numbers on Reader Service Card. Requests will be handled promptly.

Electronic Parts and Supplies 400

A 616-page 1971 catalog, No. 710, of electronics parts and supplies lists over 70,000 separate stock items from more than 700 manufacturers. Detailed specifications, descriptions and illustrations, as well as pricing cover a vast array of parts and components including-semiconductors, integrated circuit devices, tubes, relays, timers, transformers, resistors, capacitors, connectors, coils, chokes, sockets, plugs, jacks, switches, fuses, batteries, clips, lamps, wire and cable, and much more. Other major sections of the catalog feature test instruments, recording equipment, sound equipment, intercoms and other business communications gear, power supplies, electronic counters, industrial silicones, epoxy material and other chemicals, hardware, technical books, tools and solder equipment. Allied Electronics.

Wire Cable and Tubing 401

A 72-page catalog of wire, cable and tubing products, No. W-7, describes more than 7000 products manufactured and marketed by the company. It includes descriptions and specifications for products ranging from hook-up wire and multiconductor cable to coaxial cable and zipper tubing. Among the new products listed in the catalog are heat shrinkable soldering sleeves, solid and miniature Teflon wire and machine tool wire. A two-page construction index provides the reader with a quick reference for locating the various gauges and combinations of cables. Alpha.

Sound Systems

Techniques and equipment for sound systems in the sports and entertainment fields is covered in a 12page brochure. The publication illustrates the equipment used in sports arenas, stadiums, automobile speedways, hotels, restaurants and other public entertainment buildings. Altec Lansing Div.

Switch Catalog

Information on enclosed push-button and rotary switches and termination hardware is contained in the G-306-A catalog. Among the products included are: decorative push-button

switches (momentary and alternate action), environmentally sealed pushbutton switches, key-operated rotary switches, a build-your-own rotary switch kit, spring return rotary switches, and isolated position (pullto-turn or push-to-turn) rotary switches. The engineering data section of this catalog provides information on switch parameters. A push-button switch selector chart and a rotary switch selector chart allow you to determine readily the switch for your application. Another feature of the catalog is a shortform listing of the manufacturer's switches. This catalog allows you to pick a qualified switch from a variety of sizes and current ratings for rotary switches. Grayhill,

Capacitor Replacement Guide 404

A 24-page replacement guide for electrolytic capacitors is being offered by the manufacturer. It lists original part numbers for color and B/W TV sets, and includes cross reference guides for other manufacturers' products. Loral Distributor Products.

1971 Electronics Catalog 405

The 1971 edition of the Industrial Electronic Components catalog, No. FR-71-I, containing nearly 1700 items, has been released. The 84-page catalog has been revised and expanded. It lists such diverse hardware items as connectors, adapters, alignment tools, clips, plugs and jacks, binding posts, cable clamps, and cable ties. A broad selection of Nylon and metal mounting hardware, printed circuit materials, printed circuit connectors, spacers, switches, test prods, production-type wire strippers, grommets and lacing cords are included. A special appendix in the catalog provides part-number cross-references to identify parts made by other suppliers. GC Electronics.

Tools

402

403

406

407

A 100-page tool supply catalog, No. 1170, features more than 375 quality items such as cutting tools, hard-to-get taps and dies, thread plugs and ring gauges, drills, gauges, micrometers and many more. Rutland Tool & Supply.

Instrument Rental

The 42-page catalog, GEC-155D, 1971 edition, lists more than 100 new items giving the monthly rates and ordering information for a wide variety of analytical instruments, electro-mechanical measuring devices, and electrical and electronic instrumentation available to industry through rentals. Technicians are available to operate the rental instruments or to conduct electro-mechanical measurements on site. General Electric.

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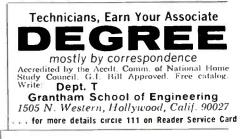
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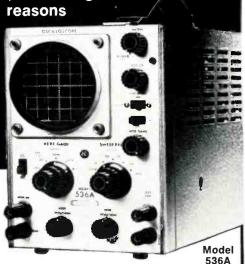
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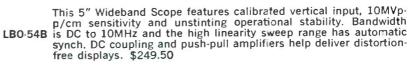
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