

Includes 3 Sections 1. VIDEO SPEED SERVICING SYSTEMS

2. TV FIELD SERVICE DATA SHEETS

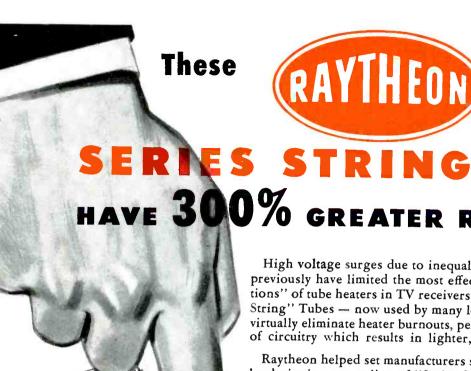
3. COMPLETE TV SERVICE INFORMATION SHEETS



The Professional Radio-Tyman's Magazine

Reaching Every Radio TV Service Firm Owner in the U.S.A.





High voltage surges due to inequalities of heater warmup time previously have limited the most effective use of "series connections" of tube heaters in TV receivers. The new Raytheon "Series String" Tubes - now used by many leading set manufacturers virtually eliminate heater burnouts, permitting the use of this type of circuitry which results in lighter, more compact receivers.

Raytheon helped set manufacturers solve this warmup problem, by designing a new line of "Series String" Tubes which feature tightened controls on heater warmup, identical current value and a heater stability so improved that heater burnouts from warmup surges are rare. By narrowing the tolerances on heater wire to one-third of the former specifications and improving heater coating techniques this has been achieved. This important advance plus Raytheon's thorough knowledge of every aspect of tube construction guarantees the superior quality of Raytheon "Series String" Tubes.

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OLD LIMITS

DEW LIMITS

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is a heater-cathode type, sharp cutoff pentode of miniature construction designed for service as a high-fre-quency amplifier in radio and television receivers.

RAYTHEON 3BC5

is a heater-cathode type-sharp cutoff pentode, of min-iature construction. Used as an RF amplifier and as a high-frequency, intermediate am-

RAYTHEON 3BN6

is a 7-pin miniature, heater-cathode type, sharp cutoff cathode type, sharp cutoff pentode. Designed to perform the combined functions of limiting and frequency dis-crimination in FM and TV receivers.

RAYTHEON 3CB6

sharp cutoff pentode of min-iature construction designed for use as an intermediate frequency amplifier, operating at frequencies in the order of 40 megacycles, or as an RF amplifier in VHF Television Tuners

RAYTHEON 5AM8

is a diode pentode of miniature construction designed for use as a video detector and IF amplifier in television re-ceivers.

RAYTHEON 5ANS

is a medium-mu triode and a sharp cutoff pentode of min-iature construction designed to perform combined functions of a video detector or LF amplifier and sync separator.

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is a heater-cathode type, double triode of miniature construction designed for mixer applications. mixer applications

RAYTHEON 5U8

is a heater-cathode type triode-pentode of miniature con-struction designed for use as an oscillator mixer.

RAYTHEON 654A

is a heater-cathode type medium-mu, high-perveance triode of miniature construc-tion for use as a vertical de-flection amplifier in TV re-ceivers.

RAYTHEON 6SN7GTB

is a dual triode designed for use as a combined vertical oscillator and vertical deflection amplifier in television receivers.

RAYTHEON 7AU7

is a heater-cathode type double triode of miniature construction designed for use as a resistance coupled volt-age amplifier, phase inverter, horizontal deflection oscillator or vertical deflection os-cillator-amplifier in television receivers.

RAYTHEON 12AX4GTA

is a heater-cathode type di-ode designed for use in Hori-zontal frequency damper service in television receivers.

RAYTHEON 12BH7A

is a heater-cathode type medium-mu double triode of miniature construction designed for use as a vertical deflection amplifier in television receivers amploving "Series" receivers employing "Series String" heater designs.

RAYTHEON 12BK5

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is a heater-cathode type beam pentode power ampli-fier. Generally used as an out-put tube in ac-dc receivers.

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is a heater-cathode type beam pentode designed for service as a vertical deflec-tion amplifier in TV receivers having a relatively low B sup-ply voltage.

Ask your Raytheon Tube Distributor about these and other new Raytheon

"Series String" Tubes

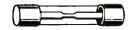
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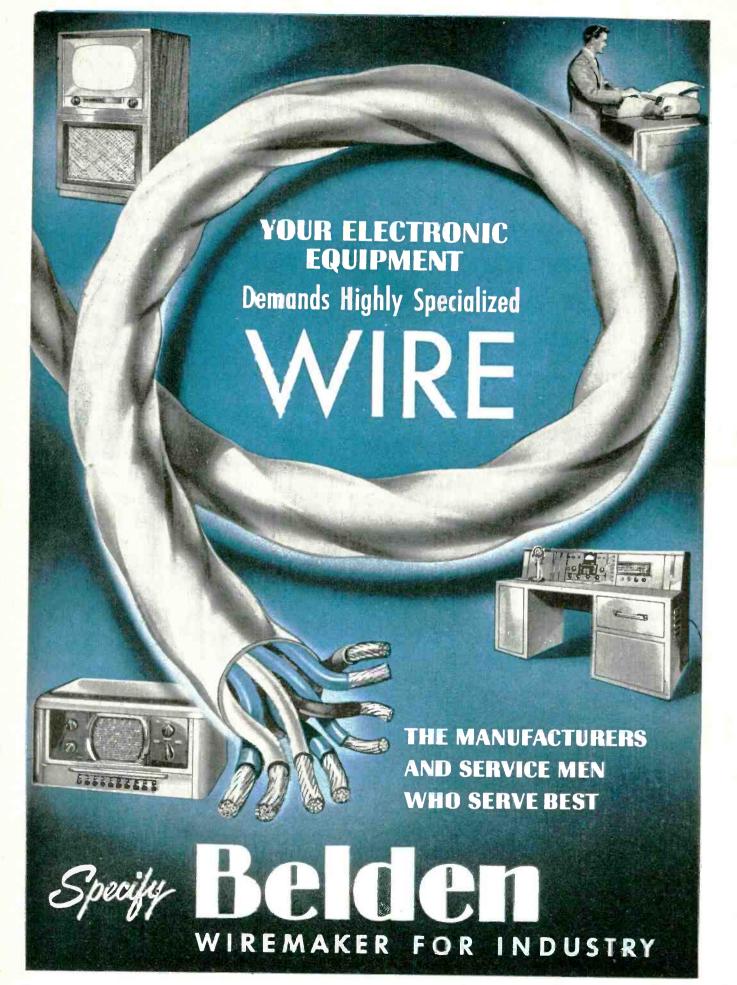
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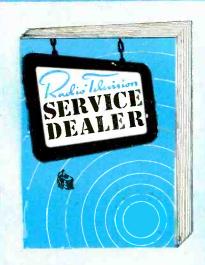
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FEATURE ARTICLES	
Discussion of signal processing of chror	ninance signal in Q and I receiver.
Understanding and Servicing Horizo Circuits, Part 2, by Oscar Fisch The structure and function of the H.O.	T. and trouble-shooting procedures.
by Walsco Electronics Corp. A unique TV receiver utilizing sectional	
TV Instrument Clinic, Part 6, by Rober Servicing techniques involving identifications	t G. Middleton 23
Transistor Portable, by Regency Commercial applications of new compo	onents and circuitry continue,
A New TV Remote Control Unit, by In Description of a versatile TV remote co	vin R. Gwirtz 28
Applications of Resistive Networks, & Applications of resistive networks in TV	by J. C. Geist
Audio Adaptations for Profitable Se Practical and profitable suggestions for	rvicing, by Gil Tint
Sync Amplifier Problems, by Paul Gold The use of the Oscilloscope in troubles	berg (a Workbench Feature) 34
Detroit Servicemen's Show, by H. Hirl	36
Distributors stage an unprecedented sh TV Service Information Sheets Complete preliminary service data for C	ow for the serviceman. 65-72 BS-Columbia Color TV Receiver, Model 205.
CIRCUIT AND SERVICE FO	RUM
Answer Man G.E. Model 24C101—Ripples in Pattern Sylvania Chassis 1-510-1—Horizontal D	27 Orifts
RCA 630 TS—Pulling at top of pix and	vertical hold 34 prizontal wobble 35 divertical slipping 35
Video Speed Servicing Systems Majestic 1218	37 121 and 122 39 41
Rider TV Field Manual Service Data S Philco, Code 140, Ch. R191, D191 Stewart Warner, Mod 9132A	43
DEPARTMENTS	
Editorial 4 Association News 8 Answer Man 27 Workbench 34	New Products 48 Trade Flashes 50 RTSD Annual Index 62 Advertising Index 64

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by S. R. COWAN PUBLISHER

Lost Opportunities-Lost Dollars

Having been associated with professional servicemen for over 25 years I fully realize that most are much more service than sales-minded. By that I mean that the average serviceman will tackle with glee any repair job regardless of the difficulty involved whereas only "exceptional" servicemen make it their standard practice to try to sell something more than mere service—things like antenna rotators, long-play needles, portable battery sets, or that second TV set for the customer's bedroom.

For example, all servicemen will repair faulty record-players or they will replace defective cartridges whenever a customer asks to have such jobs attended to. But, does the average serviceman, having finished repairing a receiver, make it his standard practice to ask his customer leading, sales-pointed questions such as: "How is your record-player working?" — or: "How about letting me install a new, more efficient TV antenna or rotator so you'll get better reception?" Sorry to say, and practically all of you will concur, the answer is an emphatic NO! The average serviceman simply allows hundreds of potential sales opportunities — and fabulous additional earnings — to slide by him month after month. Let's review this sad condition more fully!

We all know that at present the country is what one could call "phonograph record crazy." With that in mind, during the past two months I have made a thorough market study of the recordplayer business and its attendant effect upon the earnings of the service profession. My findings will undoubtedly interest all independent servicemen, and particularly service dealers and service firm owners who have one or several technicians on their payrolls. Analyze the following facts carefully:

There are more than $67\frac{1}{2}$ million record players of all types (from radio-phono combinations to portables) now in general use in this country. (As there are only 45 million homes, every other home has 2 record-players). Of the $67\frac{1}{2}$ million record-players in use, 40 million are single-speed 78 RPM; 12 million are 2-speed $33\frac{1}{3} - 78$; 2 million are 2-speed $33\frac{1}{3} - 45$; 2 million are single-speed 45; and $11\frac{1}{2}$ million are 3-speed $33\frac{1}{3} - 45 - 78$.

During this year cartridge manufacturers sold to servicemen through parts distributors for replacement use almost 1½ million cartridges. This may not sound like an impressive figure if only

given casual thought, but believe me, it is "peanuts" compared to what sales might have been had servicemen tried to get more replacement cartridge sales. Over 90% of the cartridges sold by servicemen this year were sold because the recordplayer owners gave their servicemen instructions to make the replacements. Less than 10% of the cartridges were bought by customers at the instigation of repairmen, or because servicemen "suggested" that they would get more enjoyment from their phonographs thereby.

As the public is record-player conscious, I urge all servicemen to get "sales-minded" immediately and from now on "go after" replacement cartridge sales at every opportunity — meaning whenever you have a conversation with any customer Also, when possible, ascertain if the customer wouldn't like to have his old (and probably obsolete single-play), unit replaced with a modern multi-speed turntable—a new tone arm—new long-play stylii, etc. Try this sales-minded policy in earnest for a while (and have your employed technicians do it too), and within a short time you will increase your earnings by hundreds of dollars, and probably you'll increase your technicians' earnings too, on a profit-sharing basis.

In like manner, it will pay handsomely to "go after" replacement antenna sales, remote-tuning device sales, rotator sales, and the like. If only 1 out of every 10 customers that you and your technicians have contact with becomes a buyer of some supplemental service or accessory, you will not only profit therefrom, but more important, you'll increase the customers' respect and Good Will for you as they will recognize in most cases that you acted as you did with their best interests in mind.

TV Receiver and Station Count

As of Oct. 1st, 1954 in 251 major markets there were 402 TV stations on the air, 112 of them being UHF and 286 VHF. By January 1955 an additional 29 VHF and 28 UHF stations expect to be in operation to give us a grand total of 463.

At this writing statisticians estimate that slightly over 32 million TV sets are in use of which slightly less than 1 million are UHF. An additional 3 million TV sets are on dealers' floors or are in distributors' stocks ready for delivery. There is a normal production-to-sales ratio at present, which is a good sign.



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Things are not as they seem
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Until you look inside.



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this great all-channel antenna discovery is smashing sales and performance records in every TV area!

Single bay SUPER RAINBOW model no. 331



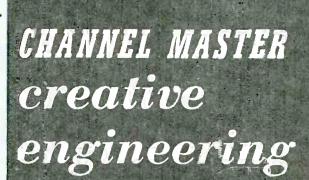
brilliant black-and-white performance—and really ready for COLOR!

these 3 revolutionary, power-packed design features — found in no other antenna today!

- 1. New spacing formula: Radical new spacing arrangements between the directors and reflectors has, for the first time, extended the full efficiency and high gain of the basic narrow band Yagi over the full width of an entire VHF band.
- 2. New "triple power" High Band directors and reflectors: Three-section directors and reflectors, with insulated segments, provide combined power of three High Band Yagis, operating side by side, in phase.
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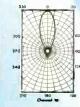


Stacked RAINBOW model no. 330-2

Here's how the RAINBOW out-performs the famous Champion:

	CHAMMEL		3	4-	. 5		,			10	11	12	
Gain Over	1-Sey RAINBOW	DB	Da	0	+1 DB	+ Z DB	+3 DB	+2.5 DB	+1 DB	+.5 DB	+.5 DB	+1.5 OB	+2. DE
1-Bay Champion	T-Bey SUPER RAINBOW	+1 DB	+1 D8	+1.5 Dil	+2.5 DB	+3.5 DB	+3.5 DB	+3 DB	+ 2 DB:	+1.5 ĐB	+ 2 BB	+3,5 DB	+4. DB
	CHAMNEL	2	3	4	5		7			10	13	12	1
Guin <u>Cver</u>	Stocked RAINBOW	+1.5 DB	+2 DB	+1.5 D8	+1.5 DB	+2 DB	+.5 DB	+.5 D8	+ 0	+O DB	+0 D#	+ 1 DB	+1.
Champion	Stucked SUPER RAIMBOW	+2	+2.5 DB	+ 3 DB	+3 DB	+4 DB	+.5 D8	+1 DB	+1 DB	+2 DB	+ Z DB	+ 2.5	+3.

horizontal polar pattern (relative voltage)



for fringe and super-fringe areas:

Super Rainbow, model no. 331 \$37⁵⁰ list stacked Super Rainbow, model no. 331-2 \$75⁷⁰ fist

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Rainbow, model no. 330 \$2360 list stacked Rainbow, model no. 330-2

\$4860 list

something new in indoor antennas



- · features 3 telescoping sections.
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- handsomely packaged for display.

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*all VHF all UHF the only indoor antenna with this "2-Way" feature.

model no. 381 \$695 list

the PRE-VU

for all-channel VHF reception only. model no. 380

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brings you today's newest installation ideas

for . . . more effective installations

- . greater customer satisfaction
- . . . higher profits for you!

TV ROTATOR

with features found in no other iotator today:

- flexible worm gear, built-in thrust bearing.
- removable motor, electrical and mechanical stops.
- weatherproof, lightweight,
- straight-thru mast mcunting, built-in chimney mount.
- extremely high torque.

ALUM ast

Aluminum Masting. The new idea in antenna masting-

can never rust!

- in telescoping sections
- in swaged 5, 10, and 14 foot sections

Lightweight ALUMast is 1/3 the weight of steel, making it so easy to install - it swings right up! Stronger than steel, ALUMast is easier to stock and actually more economical.



SELECTENNA COUPLING SYSTEM

permits unlimited antenna combinations with only one trarsmission line!

- for the first time, you can tie together an unlimited combination of antennas, including separate antennas operating on the same band.
- ideal for areas currently using rotators, manually-operated selector switches, and "omnidirectional" antennas.

list price \$542 including harcware and wire for oining couple's,

This interlocked

stace consists of 4 antenne coupiers and 1 Hi-Lo coupler; ioin: 4 antennas.

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model no. 9520, without directional indicator, \$4495 list





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Be Right With OH



DEPENDABLE RESISTANCE UNITS

ASSOCIATION NEWS

NETSDA

A meeting of NETSDA was held in the Hotel Diplomat in New York, Sunday, November 14, and attended by delegates representing associations in New York State, Pennsylvania and New Jersey. The main topic of discussion was the license bills pending in various places. Association Counsel, Mr. Joseph Forman gave an excellent review of the New York City bill with some of the problems that had to be taken care of in the formation of the bill. The necessity of having a board established which would be as impartial as possible, composed of various segments of the industry, government and the public was made apparent, and it was generally agreed by delegates present that such an impartial board to examine applicants. issue licenses and take care of future discussions between licensees and the public, would be a primary prerequisite for a good license bill.

The next meeting will be in January exact time and place to be determined

at a later date.

PRSMA-Phila.

At the October 12, 1954, open meeting, P.R.S.M.A. presented Winston Electronics, Inc., manufacturers of WinTronix test equipment. The meeting was held in the Gold Room at the Broadwood Hotel, Broad and Wood Streets, here in Phila. Winston presented three speakers—Winston H. Starks—Ralph Weinger and Daniel Kursman who spoke on color television test equipment.

Because of a change in Color Telecast, the show that was to be presented that night at 7:30 to 8:00 P.M., was not put on the air and was not seen at the meeting. P.R.S.M.A. will try again to arrange a meeting when a color telecast.

cast can be seen.

NATESA

The receipt you issue is of extreme importance. It can bind you to many things of which you may be entirely unaware and can cause you untold trouble and cost. Legally, you are bound by common law to exercise certain care over any goods in your charge.

Because of this, every service shop is the grave-yard of many junks whose owners decide they don't want to spend money on to revive. One other aspect is

[Continued on page 63]



model TR-12

A special combination value consisting of complete rotor including thrust bearing. Handsome modern cabinet with meter control dial, uses 4 wire cable.

model TR-11

The same as the TR-12 without thrust bearing, complete with meter control dial cabinet, uses 4 wire cable



model TR-2

The heavy-duty rotor with plastic cabinet featuring "Compass Control", illuminated "perfect pattern" dial, uses 8 wire cable

model TR-4

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ACHIEVEMENT OF A



For those who pursue the ultimate—the rediscovery of perspective in music...



Imperial

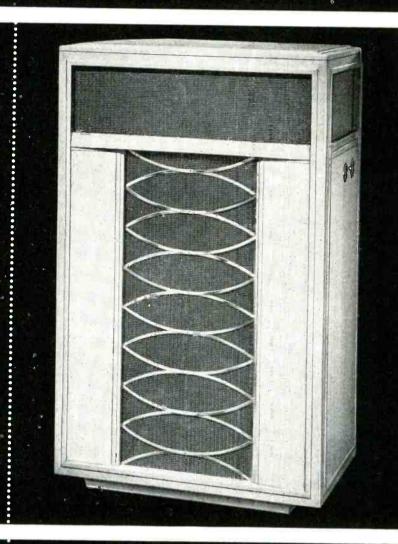
PR-100

The stimulation and pleasure gained by listening to a live performance is the result of much more than frequency range considerations.

Here is a revolution—the use of true proportions of sound in authentic reproduction including smooth coverage of the complete useful frequency range and thus recreating the fine performance with the greatest possible degree of accuracy.

Voices come to life and there's a new almost geometrical separation of instruments. A three-way system with 1-f unit loaded by a new-design reactance-annuling trilateral-mouth horn for bass; selected compression-driver horn-loaded mid channel with intrarange equalizer for a final touch to precise balance and coloration elimination; and superlatively smooth, space-blended supertweeter top. Each instrument is individually serial numbered and accompanied with a signed certificate certifying that the reproducer fully meets the exacting performance standards set for it. (Components and performance are the same as for RS-100 Laboratory Reference Standard Reproducer.)

PR-100 "IMPERIAL" REPRODUCER



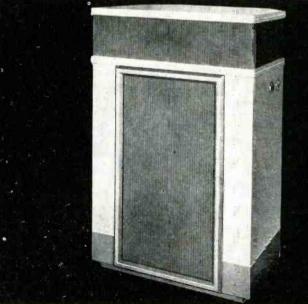
LABORATORY STANDARD

RS-100

Built for research comparison

The Imperial was designed by the Jensen engineering staff for their own use as a reference standard of the highest quality of high-fidelity reproduction. In this original laboratory version the RS-100 Laboratory Reference Standard Reproducer is a new and important tool for sound, recording and broadcast engineers, workers in psychoacoustics and music critics who require an unusually high quality of reproduction. Some music lovers and audiophiles will undoubtedly want to own an RS-100. Cabinet is plywood attractively two-toned in blue gray.

RS-100 LABORATORY REFERENCE STANDARD REPRODUCER
ST-920. Net Price.....\$468.00





Jensen-world's quality standard for more than a quarter century.

CHROMINANCE Systems

in Color TV Receivers

by

BOB DARGAN
and SAM MARSHALL

Discussion of signal processing of chrominance signal in Q and I receiver.

Part 1

THE chrominance section of a color TV receiver is used to recover and and process color signals from the composite video signal. There are two types of chrominance systems used. These are the I & Q or "Ceiling Performance" and the R-Y and B-Y or "Economy" systems. The manner in which these two systems operate will now be considered from a general descriptive point of view. A more detailed analysis will follow subsequently.

Ceiling Performance and Economy Systems

In a previous chapter entitled, "Block Diagram Analysis of Color Transmission and Reception," the entire discussion was based on the I and Q system which provides for a 1.5 mc video frequency generation of certain colors which the eye can differentiate in fine detail and a 0.5 mc video frequency generation of other colors which the eye cannot differentiate in fine detail. Such a system provides the finest available in color reception, and a receiver designed around this system is commonly referred to as a "Ceiling Performance" receiver.

It is entirely possible to extract the color difference signals without making use of the individual I and Q signals. By doing this we lose the high definition color video frequencies from zero to 0.5 mc. Such a receiver, because of the fewer circuits required is called an "Economy Receiver".

A discussion of a general block diagram encompassing both systems now follows. It must be borne in mind that the diagram shown is typical of receivers in general and does not refer to any specific type.

On examining this block diagram, shown in Fig. 1, it will be observed that the video detector feeds the demodulat-

ed video if composite signal into the section marked composite video amplifier. Here the various signal components making up the composite video signal are extracted by means of traps and filters and fed to their respective sections for further processing. These signal components are the luminance signal, the horizontal and vertical sync

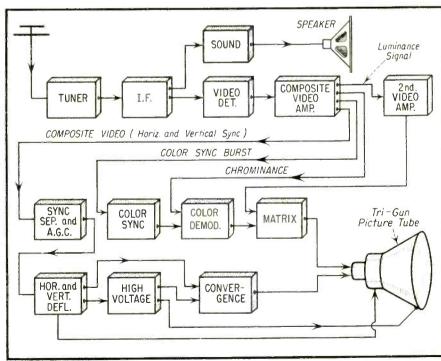


Fig. 1—A simplified block diagram of a general color TV receiver

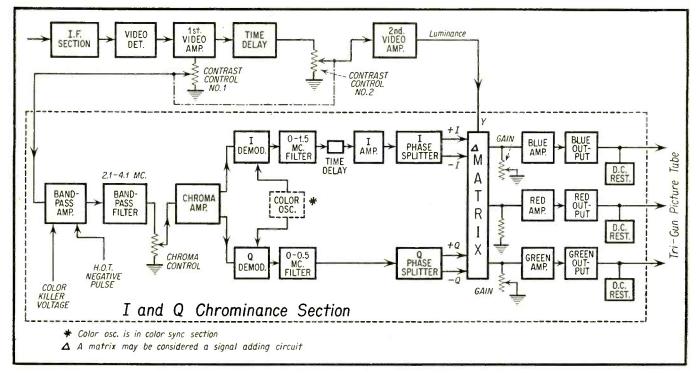


Fig. 2—The stages comprising the I and Q chrominance section.

pulses, the color sync burst and the chrominance signal.

Notice that the Y or luminance signal, which is equivalent to the black and white brightness information of the scene, is processed in a conventional manner through a 2nd video amplifier. From this point it is applied to the picture tube through the signal mixing network called the matrix. For a B & W transmission the signal in the luminance channel consists only of the black and white video signal and the standard B & W horizontal and vertical sync pulses. For a color transmission the luminance channel contains the Y signal and the horizontal and vertical color sync pulses.

It can be seen that many of the sections are common to black and white (B & W) receivers. In general, there are four sections that actually have been added with the color picture tube to bring about a change to color television. These are a Color Sync section, a Chrominance section, a Matrix section and a Convergence section.

Briefly explained, the Color Sync section processes the color burst signal so that correctly phased 3.58 mc signals are developed and fed into the color demodulators.

The demodulators into which are fed the chrominance suppressed carrier color sideband signals, extract the I and Q or R-Y and B-Y signals from the chrominance signal and apply them to the Matrix section. The Matrix section mixes the luminance and I and Q (or R-Y and B-Y) signals so that the original primary colors are presented

to the corresponding guns of the picture tube. The convergence section provides a means of producing color registration in the 3-gun shadow mask type of color picture tube.

In a conventional ceiling performance receiver, the color demodulators extract the I and Q signals from the chrominance signal. Following this the I and Q signals are fed into the matrix network where they are mixed with the Y signal to produce the primary red, green and blue signals.

In comparison, the economy receiver extracts the R-Y and B-Y signals from the chrominance signal. Following this, the R-Y and B-Y signals are mixed to produce the G-Y signal. Then all three color-difference signals are mixed or matrixed with the Y signal in a separate matrix, or directly in the picture tube via the grid and cathode terminals. The net result is the reproduction of the original primary red, green and blue signals. In the economy receiver it must be borne in mind that the processing of the chrominance signal to produce the R-Y and B-Y signals involves losing the 0.5 mc to 1.5 mc color sidebands signals. This will be explained in greater detail subsequently.

The I & Q Chrominance Section

Let us examine the stages that comprise the Chrominance section of a ceiling performance (I & Q) receiver. These are shown in Fig. 2. The color sideband signals applied to the chrominance section is obtained at the contrast potentiometer in the cathode of the video amplifier circuit. In certain re-

ceiver designs this contrast control may be located in a cathode follower stage so as to provide a low impedance source, thereby insuring less interference as the signal is coupled into the following stage.

To refresh the reader's memory it will be recalled (Fig. 3) that the detected video output contains luminance signals from zero to 4.1 mc, and a chrominance signal which extends from 2.1 to 4.1 mc around a 3.58 mc suppressed sub-carrier. This chrominance signal contains the double sideband information (from 0 to 0.5 mc) of both the "I" and "Q" signals and the single sideband information (from 0.5 mc to 1.5 mc) of the "I" signal.

The Bandpass Amplifier

Since signal information other than these chrominance signal frequencies might give rise to spurious circuit operation, that is noise signals at frequencies below 2.1 mc and crosstalk between the luminance and chrominance signals, it is desirable to suppress all other frequencies than those between 2.1 mc and 4.1 mc. For this reason the signal obtained at the contrast control #1 is coupled to a frequency limiting bandpass amplifier as shown in Fig. 2. In the plate circuit of this amplifier a filter is employed which restricts the output to a bandwidth of from 2.1 mc to 4.1 mc. See Fig. 4.

The bandpass amplifier is used to perform another function, which is the removal of the eight cycles of color burst. This is accomplished by blocking the bandpass amplifier tube for the pe-

riod corresponding to the color burst. Blocking takes place by feeding a negative spike of the horizontal sweep frequency from the horizontal output transformer. This voltage spike cuts off the bandpass amplifier during the period of the color burst, thus preventing these pulses from being present in the bandpass amplifier. If the color burst signal were to pass through this stage the overall background of the color picture would be disturbed because of the presence of the recovered burst envelope in the blue output signal. This recovered burst has an amplitude equal to or greater than the horizontal sync signal and could easily cause the blue dc restorer to set up an excessive dc component.

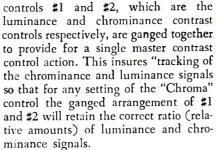
Another signal fed to the bandpass amplifier is a color killer voltage which is derived from the color sync section. (See Fig. 2). This killer voltage causes the bandpass amplifier tube to be cut off when black and white signals are being received, thus preventing signals of any sort being present at the input of the demodulator during black and white programs. If the amplifier were not made inoperative at this time it would be possible for high frequency luminance and noise signals to pass through and eventually be applied as signals to the color grids thereby causing the appearance of "confetti" on the pix tube screen. Incidentally, some receivers apply the color killer voltage to the demodulator stages.

Chroma Control

Following the bandpass filter is a chroma control which permits adjustment of the intensity of the color signals entering the I and Q demodulators. This, in turn, results in an increase or decrease of the color signal voltages applied at the input to the matrix.

Also fed into the input of the matrix is the Y or luminance signal which adds luminance information to the above mentioned color signal. The amount of this added luminance may be controlled by the contrast control #2, Fig. 2.

It will be noted in Fig. 2 that contrast



In the actual adjustment of a color TV receiver, for proper color and luminance, the sequence is as follows:

- 1. The chroma control is turned off.
- 2. The contrast control is turned off.
- 3. The brightness control is adjusted so that the screen is just below the point of illumination.
- 4. The contrast control is now turned up for the most pleasing black and white picture. Also the brightness control is readjusted to give a proper black.
- 5. The chroma control is now turned up until correct color saturation is obtained.
- 6. The Phase control is adjusted for proper hue by adjusting the phase control against known colors (such as skintones) until these colors appear as they should on the screen.

An analysis of these six steps points up the fact that the contrast control increases or decreases both the color and luminance signal at the same time. To increase the ratio of color signal to luminance, the chroma signal must be turned up. Conversely to decrease the ratio of color signal to luminance, the chroma signal must be turned down. Thus, the chroma control provides a means of controlling the color tints or saturation.

If the pre-set controls are properly adjusted in the chrominance section (particularly in the matrix and color amplifier gain circuits) then, varying the chroma control (see Fig. 2) should produce a variation of only the relative saturation of the color voltages applied to the picture tube and not the colors themselves. In fact, a check on the proper adjustment of the matrix and individual color gain controls is to vary the chroma

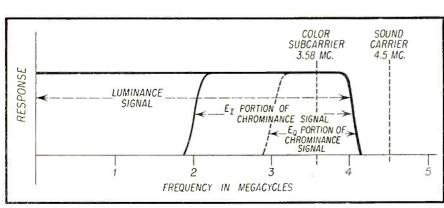


Fig. 3—Complete color picture signal pass band.

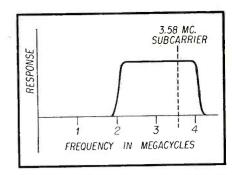


Fig. 4—Bandpass response of bandpass filter in output of bandpass amplifier.

control carefully and to observe the effect on the color picture tube screen. If these variations result in a change in color or hue the previously mentioned matrix and gain control adjustments are at fault. On the other hand, if varying the chroma control as outlined produces a change in color saturation only, the matrix and gain control adjustments are correct.

Chroma Amplifier

Following the bandpass amplifier and its filter network we usually find an additional stage known as the chroma amplifier. This stage increases the peak to peak amplitude of the color signal at the output of the chroma control and provides equal chrominance signal voltages to both the I and Q demodulators.

I & Q Demodulators

Before analyzing the action taking place in a demodulator it would be well to recall the original make-up of the chrominance signal. It will be recalled that the Q signal is made up of low video frequencies (0 to 0.5 mc) corresponding to the low acuity colors around purple and green and the I signal is made up of the high video frequencies (0 to 1.5 mc) corresponding to the high acuity colors around orange and cyan. It will also be recalled that the I and Q signals modulate a set of 3.58 mc quadrature carriers and are transmitted on a single suppressed carrier frequency as a pair of signals displaced by a phase angle of 90 degrees. The suppressed carrier Q signal is double sidebanded and the suppressed carrier I signal is double sidebanded up to 0.5 mc on each side of the carrier and single sidebanded from 0.5 mc to 1.5 mc on one side of the carrier. These two signals comprise the chrominance signal.

The primary purpose of the demodulator is to separate the I and Q suppressed carrier sidebands combined in the chrominance signal, and to reconvert them back into video frequencies identical to the original I and Q signals. To do this the input of each demodulator is fed the chrominance signal and



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a separate 3.58 mc oscillator signal, the latter being applied as follows: The phase of the first 3.58 mc signal fed to the I demodulator is in phase with the I component of the suppressed carrier chrominance signal, and the phase of the second 3.58 mc fed to the Q demodulator is in phase with the Q component of the suppressed carrier chrominance signal.

The plate circuits of these demodulators, under the above mentioned conditions contain signal products, which, in the case of the I demodulator is the I signal in its original video frequency form. The manner in which these transitions take place will be described in

greater detail subsequently.

It might be pointed out that the action described above is strictly true for the I demodulator but not for the Q demodulator. The reason for this is as follows:

In the I demodulator the complete I signal is passed by the circuit because it is in phase with the first 3.58 mc oscillator signal applied. The Q signal being 90 degrees out of phase with the first 3.58 mc signal develops equal and opposite voltages in the plate circuit which cancel out because it is double sidebanded; so that the Q signal is rejected.

In the Q demodulator the complete Q signal is passed by the circuit because it is in phase with the second or quadrature 3.58 mc signal applied. The I signal, however, is rejected only for that portion of its spectrum where it is double sidebanded (0 to 0.5 mc) for the same reason that the double sidebanded Q signal is rejected in the I demodulator. From 0.5 mc to 1.5 mc the I signal is single sidebanded, and that portion of the I signal is not rejected in the Q demodulator. This component of I signal is referred to as "Quadrature Crosstalk," and must be eliminated. This is done by the insertion of a 0 to 0.5 mc filter in the output of the O demodulator which effectively rejects the 0.5 to 1.5 mc quadrature crosstalk components previously ferred to.

Referring to the block diagram of Fig. 2 it will be observed that the I demodulator contains a 0 to 1.5 mc filter. This filter in the I channel is primarily designed to provide attenuation of signals other than the 0 to 1.5 mc video color signals which may find their way through the demodulators. Such signals are derived from the luminance channel.

Time Delay Networks

The passage of the Q signal through the zero to 0.5 mc filter introduces a time delay which is somewhat greater [Continued on page 54]



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PART 2

A discussion of the operating principles of horizontal output transformers, together with procedures designed to facilitate rapid servicing and replacement.

IN the previous installment the impedance matching pedance matching function of the horizontal output transformer was discussed in some detail. It was pointed out that the matching of the impedance of the horizontal deflection coils (the load) to that of the horizontal output tube (the generator), is a necessary requirement if maximum power transfer is to be obtained.

This may all be interesting from a theoretical point of view, but it might be worth while at this point to examine the practical implications of impedance matching. Once again, a very close comparison can be made between the corresponding parts of a radio receiver and a TV receiver. In the radio receiver, the action taking place from the audio

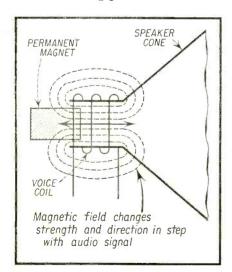


Fig. IA—Action of magnetic field in speaker.

output tube to the speaker may be summarized as follows:

1. The audio output tube (see Fig. 1A) delivers power to the output transformer in the form of an audio signal

2. The output transformer delivers this power to the voice coil of the

3. The audio signal current through the voice coil produces a magnetic field, the strength of which varies in step with the strength of the current.

4. This varying magnetic field reacts with the steady field of the permanent magnet (or the electromagnet of an electro magnetic speaker). The reaction between these two fields results in a movement of the speaker cone and the consequent production of sound.

5. The greater the variations in the magnetic field through the voice coil, the louder will be the sound issuing from the speaker.

6. Thus, to get the loudest sound from the speaker, requires the delivery maximum power to the voice coil from the output tube. The output transformer provides this maximum possible power by matching the voice coil impedance to that of the output tube.

A step by step comparison with the action in a TV receiver shows a very close correspondence, as follows, and as illustrated by Fig. 1B.

1. The horizontal output tube delivers power to the horizontal output transformer in the form of a sawtooth horizontal sweep current.

2. The output transformer delivers this power to the horizontal deflection coils in the deflection yoke.

3. The sweep signal current, flowing through the horizontal deflection coils, produces a magnetic field through the neck of the picture tube. The strength of this field varies in step with the strength of the current.

4. This magnetic field is directly across the path of the electron beam within the cathode ray tube. As the beam passes through the field, it is deflected to the left or to the right, depending on the polarity of the field. The amount of deflection depends on strength of the magnetic field.

5. The greater the variation in magnetic field strength, the greater will be the deflection of the electron beam.

6. Thus, to produce the maximum deflection of the electron beam requires the delivery of a maximum current swing to the deflection coils. This in turn produces a maximum overall change in magnetic field strength. The horizontal output transformer provides for the delivery of maximum available power by matching the impedance of

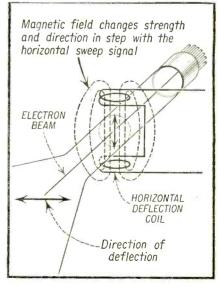


Fig. IB—Action of magnetic field in horizontal defl. coils of CRT.

the deflection coil to that of the hori-

zontal output tube.

How does all this bear on the problem a service technician faces when the horizontal output transformer must be placed? Actually, it must be admitted hat what nearly always happens is that the technician hies himself to his favorite distributor and says "Hiya Jerry! I need a flyback for an Admiral 121 M 12A." Jerry then thumbs through the pages of a replacement catalog and comes up with a Stancor A-8134 or other make replacement flyback transformer. Another common procedure would be to get an exact replacement from the manufacturer. Neither of these procedures requires any extensive knowledge on the part of the technician, so why bother with all this business of impedance matching, deflection angles, etc., etc. Our answer is that when a serviceman understands these basic principles, he becomes more proficient in diagnosing and remedying troubles. This increased proficiency can be translated directly into dollars and cents by cutting down substantially on the time required for

In the light of the previous discussion, we might summarize the deflection power requirements as follows. Suppose a receiver uses a 19GP4 picture tube. Examining the characteristics of this tube shows that the deflection angle is 66°. This is the starting point for which the other components in the deflection circuit are designed. In Fig. 2, the line OA represents the electron beam of the left end of the raster, while OB is its position at the right hand end. The deflection angle of a CRT, in this case 66°, is the angle formed by these two extreme positions of the beam.

The force which is used to deflect the beam over this angle is derived from the saw tooth wave of current flowing through the deflection coils, as mentioned previously. Fig. 3 is a graph of this current. A and B on the solid line graph are the maximum negative and positive values of the current through the coils, and at these current values the magnetic field deflects the beam to OA and OB respectively, in Fig. 2.

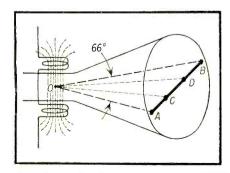


Fig. 2—Deflection angle of a CRT (19GP4).

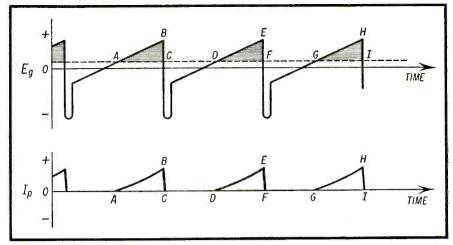


Fig. 4B—Signal voltage at grid of horizontal output tube. Current flows only during shaded portion of signal. Fig. 4C—Current flow in the plate circuit of the horizontal output tube.

The next step is to design the deflection coils so that the desired deflection may be produced with the available power from the output stage. Thus we have 70° yokes, 57° yokes, etc. If, for any reason, the current through the deflection coils is reduced, as shown by the dotted line of Fig. 3, the beam will be swept through a smaller angle. This would be the angle formed by CO and DO of Fig. 2, and of course indicates reduced width of the picture.

Obviously, an improper match could be responsible for such a reduction in sweep width. The properly designed horizontal output transformer, then forms the final connecting link in the chain between the horizontal output tube and the electron beam in the CRT.

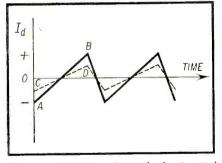


Fig. 3—Current through horizontal deflection coils, corresponding to deflections in Fig. 2.

This is why for example, flyback transformer and deflection yokes are frequently sold as matched sets for larger tube conversion jobs. This is also why the technician, when replacing either the flyback transformer or the deflection yoke must always consider whether the replacement part will provide proper matching.

Replacement transformers of the "universal" type are designed for angular deflections of from 50° to 70°. Any deflection angle within this range may be

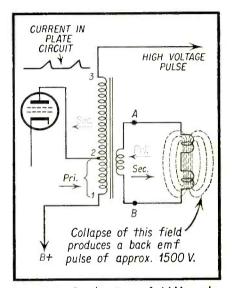


Fig. 4A—Production of H.V. pulse for 2nd anode.

handled by selecting the proper taps. For 24 inch picture tubes and larger, 90° deflection angles are used. The universal replacement transformer, therefore cannot be used in this application. Flyback transformers specifically designed for a 90° deflection must be used.

High Voltage Production

It was pointed out in the previous installment that the production of a high voltage for the picture tube second anode is another major function of the horizontal output transformer. This must be taken with a very slight grain of salt. There are a few receivers (very few) which use an rf power supply for high voltage production. These receivers use the basic horizontal output transformer, which is not concerned with high voltage production. For clarity, this is the only type we might truly call a horizontal output transformer.

The much more common type which [Continued on page 56]

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3. Remote control operation.

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Fig. 1—Tube and printed circuit strip layout of Walsco Model PC9 TV receiver. Shown also are the various adjustment controls available for immediate adjustment.

O68Q7

TUBE & STRIP LAYOUT CHASSIS MODEL PC9 portant test points are at the technician's fingertips for quick testing and diagnosis. Finally, in cases where alignment adjustments are necessary, they too are immediately accessible.

Of even greater significance than the above, from the point of view of servicing, is the possible technique of stocking replacement strips and providing quick and immediate service by installing a complete strip for one in which any of the components is defective or misaligned. In this manner the customer receives a "factory-rebuilt" replacement, a technique which is used today in many fields for quicker service and greater

The chassis design, especially with the remote control feature, lends itself to custom installations and cabinet design, with a flexibility that should satisfy the wildest-eyed interior decorator. Here again is an item the dealer or technician can sell with considerable latitude for the inclusion of accessories and installation charges.

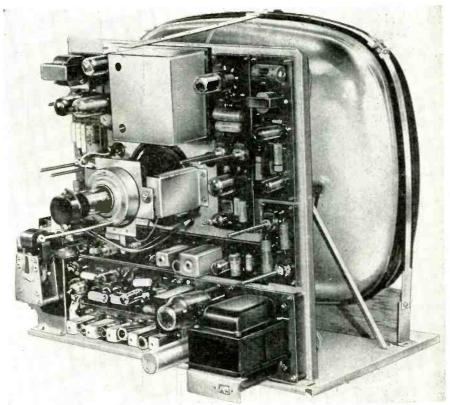


Fig. 2—Appearance of receiver on removal of rear panel.

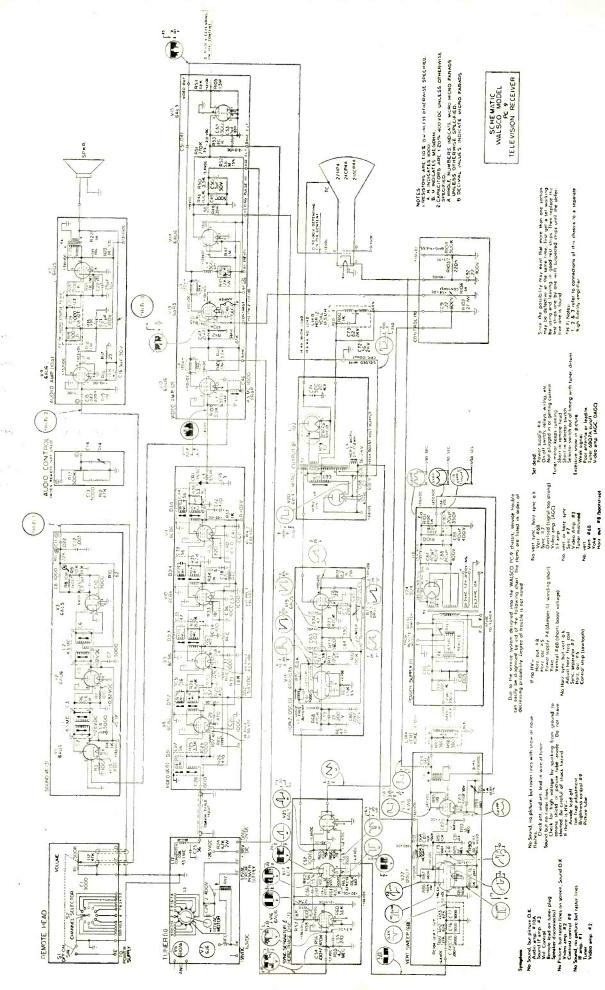
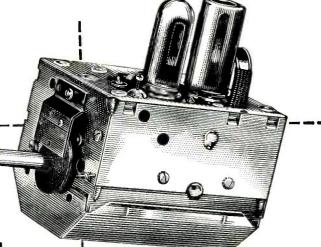


Fig. 3—Circuit diagram of Walsco, Model PC9 printed circuits receiver.

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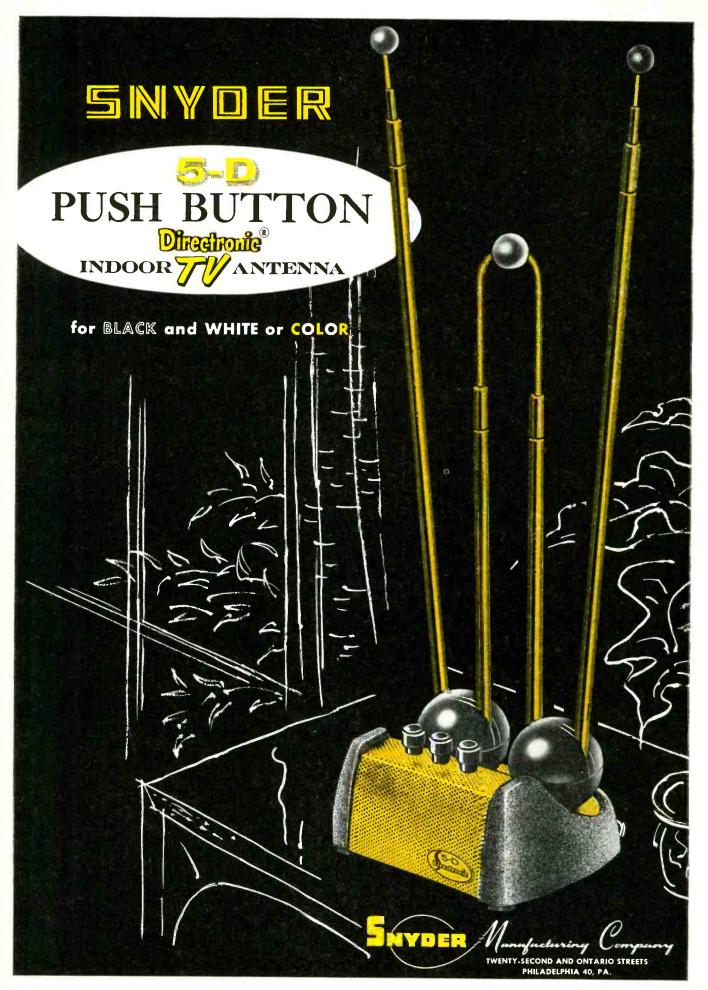
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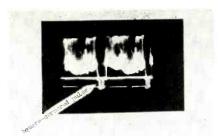


Fig. 1—Scopes with wide band vertical amplifiers display square corners on pulses.

- Q. I have two scopes in the shop, but they do not show the same waveforms when testing horizontal sync pulses; also, when testing vertical sync pulses, the horizontal lines in the pattern appear much brighter than the vertical lines in the pattern. What is the reason for these troubles?
- A. Scopes with wide-band vertical amplifiers display the square corners of pulses, as shown in Fig. 1. However, scopes with narrow-band vertical

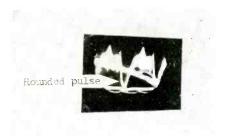


Fig. 2—Rounded pulse display.

amplifiers round off the corners of pulses, as illustrated in Fig. 2. Waveform displays often exhibit considerable difference in the brightness of horizontal and vertical excursions in the pattern (Fig. 3) because of velocity modulation; i.e., the beam may

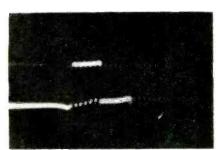


Fig. 3 — Difference in brightness caused by velocity modulation.

be traveling faster in the vertical direction than in the horizontal direction; this difference in the speed of beam travel results in velocity modulation of the pattern brightness. Velocity modulation becomes less apparent when the vertical-gain control

TV INSTRUMENT CLINIC

PART 5

Based on CHALLENGE CLINIC demonstrations, this new series discusses many measurement and test problems raised by service technicians.

By ROBERT G. MIDDLETON

Field Engineer, Simpson Electric Co.

is reduced in setting, and the horizontal-gain control is advanced in setting. To avoid distortion of high-frequency waveforms such as horizontal sync pulses, make direct connection of the video-amplifier output to the vertical-deflection plates to the CRT in the scope.

- Q. When checking the video signal in some receivers, the vertical sync pulse appears to fall below the level of the horizontal pulses. Then, after the vertical pulse passes, the response rises again. What causes this trouble?
- A. The situation noted above is illustrated in the photo of Fig. 4B. The vertical sync pulse falls in the display because of differentiating action in the video amplifier. For example, the coupling capacitor C shown in Fig. 4A may be found to be too small, or the grid-leak resistor R may be found to be too small. As a result, the theoretical signal shown in the diagram becomes distorted in practice as seen in the photo. (Note that the scope display is "upside down" with respect to the diagram.)
- Q. My scope does not sync tightly on some of the waveforms encountered in the TV chassis. What can be done to improve the sync lock?

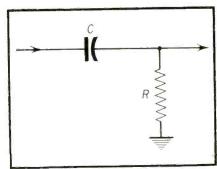


Fig 4A—An RC coupling arrangement such as used in a typical video amplifier stage.

A. A sync-sharpener may be used, as shown in Fig. 5. The sync-sharpener differentiates the waveform, and converts it into spikes which are usually more effective in locking the horizontal sweep circuit of the scope. Video signals are often difficult to lock because the camera signal is a random changing voltage; in such case, the

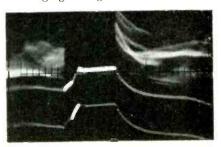


Fig. 4B—When the time constant RC is too small to accommodate the vertical blanking pedestal, the pedestal becomes differentiated (low frequencies are lost and shifted in phase).

operator will often find it advantageous to utilize the external sync facility of the scope. Switch the scope

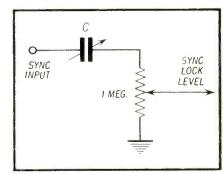
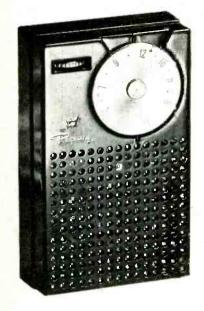


Fig. 5—A sync-sharpener arrangement for obtaining better sync lock of the pattern on the scope screen, under difficult circumstances. The value of C must be determined experimentally.

to external sync, and run a test lead from the external sync post to a point within a few inches of the picture tube screen. The pulse radiation from the picture tube will provide a clean tight locking voltage.

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The commercial applications of new components and circuitry continues without a letup as is evidenced by this revolutionary portable radio

Full view of transistor portable radio. Relative size compared to that of a teacup is shown in the upper right corner.

I.I. T.2 Ferrite Core SW-1 Loop Antenna Off-On Switch Earphone Jack C2. C3 Tuning Condenser Volume Control C9 Electrolytic X2 I.F. Transistor Dl Crystal Diode TI-221 T3 Output I. F. Transformer Tl Input I. F. C19 Transformer Electrolytic T2 Interstage K3 I. F. Transistor I. F. Transformer T1-221 L3, L4 Osc. Coil X4 Audio Transistor Xl Converter TI-210 Transistor T1-223 4 Audio Output Transformer Speaker C21 CIT Electrolytic Electrolytic Battery Regency Type 216

Exposed view of chassis shows various components corresponding to circuit diagram notations. Observe use of miniature components.

THE revolutionary pocket radio described below is the first in the history of the electronics industry to use transistors instead of tubes, and is now in production by Regency of Indianapolis.

According to Mr. Edward C. Tudor, I.D.E.A. prexy, the Regency pocket radio represents the first commercial adaptation of the transistor to a radio. "We were able to achieve this—and the accompanying miniaturization—through close cooperation with Texas Instruments Incorporated of Dallas, who effectively engineered and produced the essential high performance transistors that made our radio possible," he said.

"The development of the transistor itself marks a significant milestone in the history of electronics," he said. "Just as its perfection and adaptation enabled us to manufacture the Regency pocket radio, so does its development promise to change the size, appearance and nature of hundreds of other items in our daily life—possibly even our way of life itself," he added. "We are pleased to have worked with Texas Instruments, a company so prominent in this important revolution in our industry. This transistor development, in our opinion, rivals their important exclusive production of silicon transistors and assures the industry of more developments to come," he added.

Technical Features

Some of the interesting technical features of this receiver are as follows:

The power supply is a single 22½ volt hearing aid battery. The power drain is approximately 4.3 ma with no signal, reducing to about 4 ma with ave

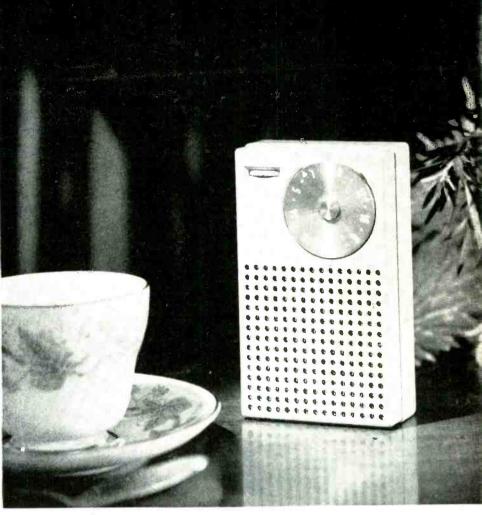


action on a local station. Battery life varies between 20 and 30 hours, depending on use.

New high performance NPN, grown junction germanium type transistors are used, developed by Texas Instruments Incorporated, for this receiver aplication. These are as follows (see schematic):

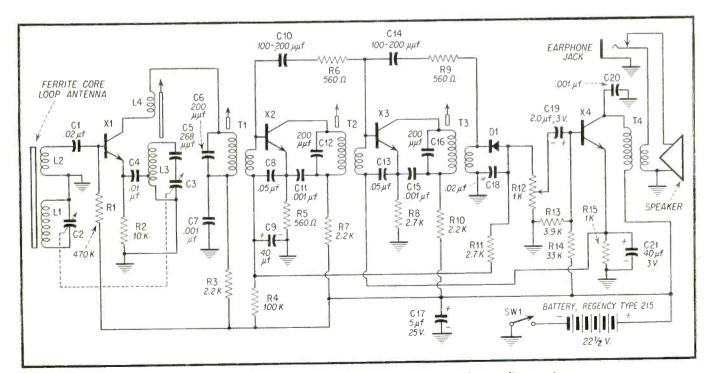
X1. Converter Oscillator—Texas Instruments No. TI-223

X2. First Intermediate Frequency Amplifier—Texas Instruments No. TI-221



- X3. Second Intermediate Frequency Amplifier—Texas Instruments No. TI-221
- X4. Output Audio Amplifier Texas Instruments No. TI-210
- D1. Is a germanium diode with a selected characteristic for low forward impedance at low signal levels

Raytheon—CK706A or Tungsol—TS117



Circuit diagram of Regency Model TR-1 transistor pocket radio receiver.

A new standard in electrical and mechanical perfection in all 32, new YAGI antennas

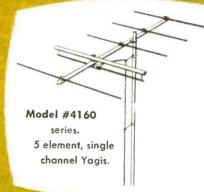


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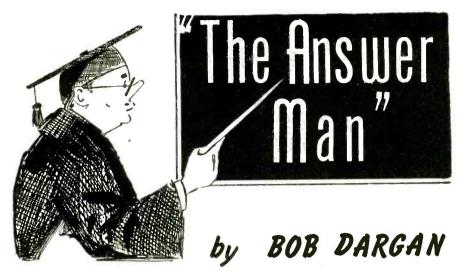
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Do you have a vexing problem on the repair of some radio or TV set? If so, send it in to the Answer Man, care of this magazine. All inquiries acknowledged and answered.

Note: Only communications with Radio-TV Service Firm letterheads will be considered and answered. Please indicate make, model, and chassis number of receiver.

G.E. Model 24C101— Ripples in Pattern

Dear Answer Man:

I have a G.E. receiver Model 24C101 that has a peculiar condition in it. There are ripples in the picture but they are not horizontal or vertical ripples as far as I can determine. In examining the condition on a test pattern they seem to be more pronounced in the outer circle of the pattern. These waves seem more noticable at the top of the test pattern circle and are present with any channel selected.

I have checked the usual parts such as tubes, voltages, resistors, etc. The tubes have all been replaced and all voltages are correct as indicated in the service literature. The only thing that might be of help is the fact that this ripple can be affected a little with the horizontal hold control adjustments. At a certain point the waviness is reduced to a minimum but still is noticeable. Whan can be causing this trouble?

E.V. Philadelphia, Pa.

The condition you have described is usually termed "Piecrusting" and is due to an open condenser in the automatic frequency control circuit of the horizontal oscillator.

The horizontal *afc* circuits provide a correction voltage to be applied to the horizontal oscillator circuit; and in this case to the reactance tube. This correc-

PULSE FROM HORIZ. OUTPUT ½12AU7 TRANSFORMER REACTANCE 4700 TUBE **6T8** .05 HOR. DISC. 150 K HORIZ. SYNC PULSE 120 1 MEG. C506 C 505 68 K 015 µf 1000 1 MEG. 1000 1 MEG. Check for OPEN Condenser

Fig. 1—G.E. Model 24C101 Horizontal Phase Comparer circuit and reactance tube.

tion voltage should bring the oscillator to the proper operating frequency. However, if there is something wrong in the afe circuit so that an incorrect voltage is applied to the oscillator or its reactance tube the circuit will hunt for the correct frequency where the sync pulse is locked in with the deflection system. With this hunting process will be a slight shift in phase of certain portions of the picture. A group of horizontal lines will be shifted first to the right and then to the left as the horizontal oscillator hunts for the correct frequency and phase. The effect is a ripple on the outer circles of a test pattern or what can easily be recognized as "Piecrusting".

The afc circuit, because of a defective component, will over-correct in one direction by applying too much voltage to the horizontal oscillator circuit. This over-correction then causes the horizontal oscillator to shift in phase a slight amount. Then it will over-correct in the opposite direction. This results in a slight shift in certain portions of the picture, each horizontal sweep line not being directly below the previous line.

Varying the horizontal hold control will change the amount of ripple in the picture because this alters the horizontal oscillator voltage in phase. Therefore, varying the horizontal hold control will permit varying degrees of oscillator hunting and different amounts of ripple.

As shown in Fig. 1 three filter condensers should be checked to determine which condenser is open. The most likely condenser to suspect is the .015 uf condenser in series with 68K resistor.

Sylvania Chassis 1-510-1— Horizontal Drifts

Dear Answer Man:

In servicing a Sylvania Chassis 1-510-1 the reason for the receiver being taken to the shop was for poor vertical lock-in action. The integrator network was checked and found to be open. Another one was obtained and installed. Then I found that I also had horizontal sync trouble. The integrating network corrected the vertical sync trouble but the horizontal oscillator drifts as though there were no sync pulses to lock it in. I have gone over the circuits a number of times and I haven't been able to locate the trouble. Do you have any ideas on this one? Could there be something wrong in the new integrator network?

D. S. St. Louis, Mo.

Undoubtedly what has hapened is that in the installation of the integrating network the terminals were reversed.

These small printed circuits are good and have been used in quite a number [Continued on page 54]



Remote Control Tuner

Los Angeles, Calif.

Description of a versatile TV remote control unit, a good source of income for the enterprising and sales-minded dealer or serviceman.

 ${f T}^{
m HE}$ advent of remote control units for television receivers has opened another profit door for the TV dealer and technician. These accessories are now priced so that all owners of TV receivers can afford them. A little enterprise on the part of the dealer in making his customers aware of the advantages of this device should net him handsome returns in the way of sales and installation fees.

The De Luxe model of the Remot-O-Matic provides a remote unit which gives the user control of six functions, these being:

- 1. On-and-off control
- 2. Sound control
- 3. Brightness control
- 4. Contrast control
- 5. Channel switching
- 6. The convenience of taking the sound control away from the set so that the sound is controlled from the remote unit or at the set itself . . . as desired.

Described below are the installation instructions for the Remot-O-Matic remote control unit. These instructions are presented in order to give the technician an idea of the relative ease with which these units can be installed, Detailed wiring instructions have been omitted inasmuch as the wiring diagram

shown on the next page is self-explanatory. However, detailed wiring instructions do come with every unit.

Installation Instructions:

- 1. Remove chassis from the cabinet and place it on working area so that the tuner is easily accessible.
- 2. The Motor Assembly is to be mounted directly behind the tuner with the shaft of the Motor Assembly lined with the shaft of the tuner as follows:
- There must be room for it . . . or the Remot-O-Matic Adapter kit must be used for interior installa-

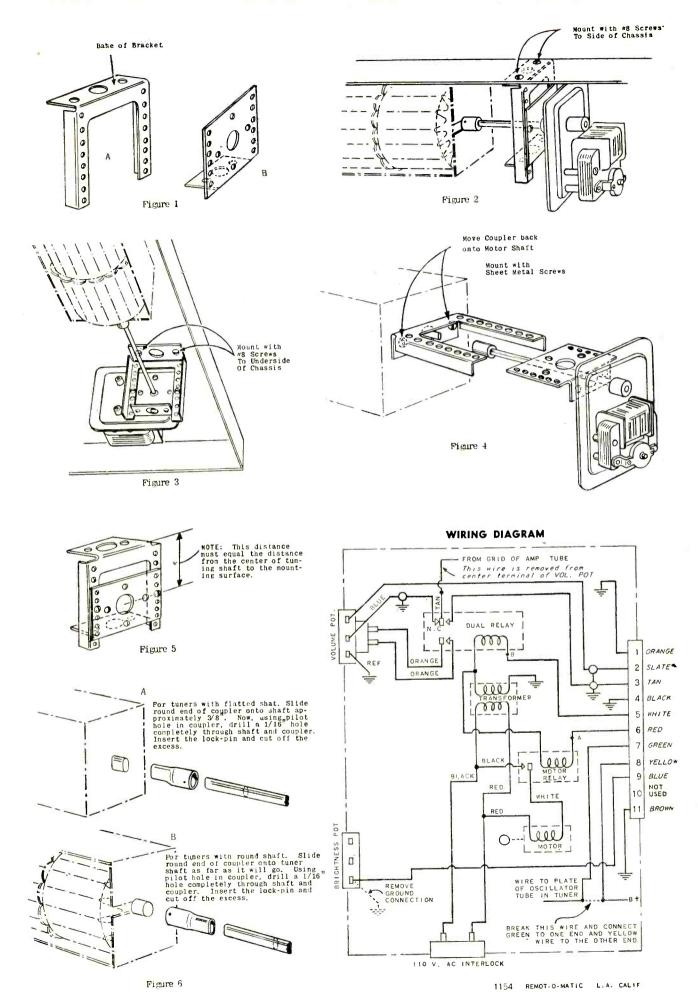
Select the location to be used for mounting the brackets (Figure 1). Various designs of chassis allow for different locations for the mounting brackets. In Figures 2, 3 and 4 you will see three different locations and methods for mounting the brackets. These three locations and methods cover nearly all possibilities, however, the brackets are so constructed that they allow for wide range of positioning in increments of 1/16".

Where the methods in Figure 2 or Figure 3 apply, take the following steps:

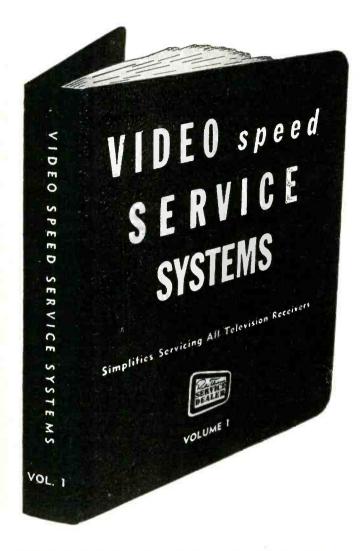
- A. Decide on the surface where you intend to mount the bracket whether it be on the side of the chassis or the main base of the chassis.
- B. Place brackets of Figure 1 together to form complete mounting brackets as shown in Figure 5.
- 4. Mount the Motor Assembly to bracket B (Figure 1) and secure with two #8 screws and lock washers . . . into the mounting studs
- of the Motor Assembly. We have a 6" shaft on the Motor Assembly which may be cut to any length. Place the Motor Assembly in the approximate position you have in the chassis and thus determine the length of shaft needed. It is important that a gap of from 1/32" to 1/8" be maintained between the shaft of the tuner and the shaft of the motor.

Note: If the tuner shaft does not protrude through the rear wall of the tuner, it will be necessary to drill a hole in the rear wall of tuner so that the coupler can slide onto the tuner shaft without binding.

[Continued on page 55]



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applications

J. C. Geist

Comprehensive design data and applications of resistive networks in TV measurements and antenna signal termination and distribution.

MOST technicians have used resistive networks in some phase of TV service work. However, there does not seem to be a general recognition of the wide variety of uses these simple devices can serve as auxiliary service aids. Table 1 lists a number of different types of resistive networks and the pertinent characteristics of each type. The table in turn refers to the applicable circuit configurations which are shown in the attached drawings. While it is true that for most of the applications listed, there are other devices available with less signal losses and more refined characteristics, the resistive network generally represents the simplest, least expensive and most convenient answer when the losses can be tolerated. Furthermore, when arranged compactly, using short lead lengths, the resistive network will be useful over extremely wide frequency ranges. Simple networks made up of standard composition resistors will give

		T	DLFI			
Notwork	Configuration	Resistor Values in Ohms	Nonunal Z1 in Ohms	Nominal Z2 in Ohms	E2)	Insertion Loss in di
A	ş	R1-110 R2'82 fr3 150	7.2	300	54	5. 3
В	t	R1 *130 R2 +56 R3 *150	50	300	52	5, 7
С	2	R1 *82 R2*220	300	300	- 12	10
D	2	R1*120 R2 • 68	300	300	10	20
E	3	R1 "82 R2*160 RJ*220	300	300	. 10	20
F	4	R1+100 R2=140	72	72	. 32	10
G ·	4	R1:360 R2:88	72	72	10	20

Table 1—Characteristics of configurations discussed in the text.

good performance over the entire range covered by the high and low television vlif channels. With sufficient care in construction using the smallest resistors and very short lead lengths this range can even be extended into the uhf television band. Some of the ways in which these networks can be advantageously put to use are discussed below.

Sweep Generator/Receiver Impedance Matching

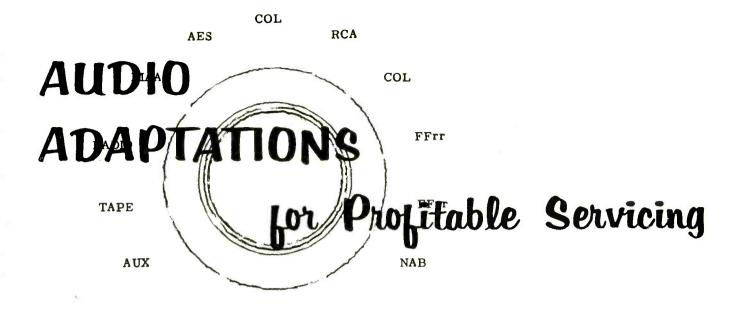
Resistive

Networks

When a sweep-frequency signal generator is used for TV tuner alignment the transmission line from the generator must be terminated in its characteristic impedance. If the termination is not of the correct value the transmission line will act as a frequency sensitive tuning element (VSWR varying with frequency) and thus cause an erroneous indication of the bandwidth characteristic of the tuner. Most sweep generators utilize a low-impedance unbalanced (one side grounded) coax output to simplify output attenuator construction and to reduce leakage. Since the tuner has a balanced (neither side grounded) antenna input with an impedance of 300 ohms it is necessary for the generator coax to be terminated by a device which will match the high receiver impedance to the low coax impedance and at the same time convert from a balanced to an unbalanced circuit. Networks A or B, depending on the generator coax impedance, are designed to do this job. The resistive network does not provide a completely balanced output as do other devices such as rf transformers or "balun" transmission line networks and it introduces greater loses than these latter devices. However, the resistive network provides an output sufficiently close to perfect balance to accomplish a satisfactory termination, and at the same time offers the advantages of extreme simplicity and usefulness over a wide frequency range. The relatively low
[Continued on page 57]

				TABLE 2		E2/E1	Insertion Loss tn db	Receiver Isolation in db
Configuration	Receiver Outlets	Resistor Values in Ohms	Nominal ZI in Ohms	Antenna Line Termination in Ohms				
5	2	H ₁ = 200 R ₂ = 680	600	600 2 = 300	300	.35	ũ	18
5	3	R ₁ = 350 R ₂ = 510	300	900 = 300 3	300	.21,	14	28
5	4	R ₁ = 550 R ₂ = 390	1200	1200 = 306 4	300	13	18	36
6	2	R ₁ = 240 R ₂ = 180 R ₃ = 32	900	300 = 300 2	300	.14	17	34

Table 2—Resistor values and characteristics for different combinations of impedance and number of receivers in the system.



by

Practical and profitable
suggestions for increasing the utility of
Hi-Fi systems

THIS article provides a number of adaptations that may be made to obtain more versatile performance of audio equipment where a hi-fi installation has already been made. The modifications outlined herein do not require chassis pulling or alteration. All work is done primarily by means of adapters and external mountings.

An effective way to satisfy the hi-ficustomer is to provide him with "extras"—features which make his installation more than just an interconnected group of ready-made components. What is nice about these modifications is that, while they not only make an audio system more attractive, they are also simple to install and add very little to the audio man's costs. Here are a few such additions which the writer has found quite useful.

Feeding TV Sound Through the HI-FI System

Almost every TV ad one sees stresses the advantages of FM sound. The average consumer is led to believe that his set is capable of wide range audio reproduction. Unhappily, this is not the case. After the sound information passes (relatively unscathed) through the tuner, if stages, and FM detector, it is squeezed through one single-ended audio stage and fed to a speaker just about up to reproducing a telephone conversation faithfully. Why waste this hi-fi potential? This is a good question to ask the customer-from a sales point of view. Invariably, he will welcome the chance to improve his TV sound by

utilizing his audio system. The necessary circuit changes could be made by pulling the chassis, drilling the required holes, mounting a terminal strip and making the electrical changes. As every technician knows, however, chassis-pulling involves extra time and labor and should be avoided where possible. The advantages of the methods described here are that all work takes place on the top side of the chassis, and the adapters can be prepared during free time and merely plugged in when needed.

First an "S" type of socket (Fig. 1, item A) is fitted into the shell of a tube base (B). The \$5 pin of the "S" socket is grounded, thereby grounding out the grid of the audio output tube. The \$5 pin of the tube base (B) connects internally to the output connection of the first audio tube so that a lead connected to this point may be brought out to the hi-fi power amplifier.

The diagram in the lower right hand corner of Fig. 1 shows how this modification renders the output stage of the TV receiver inoperative. The incoming signal, on the other hand, is transferred through the connecting cable to the desired hi-fi audio amplifier and speaker combination.

A shielded cable length of no more than 5 feet is practical in this case because of the high impedance circuit involved. Excessive hum pick-up and highfrequency attenuation would result if a longer cable were used. Aside from the cable, the only materials needed are the tube base of an old octal tube, a suitable phono-input connector and an Amphenol "S" type octal socket, 78-S8. This socket is perfect for this application since it exactly fits the tube base. A hole should be drilled in the side of the base near pin 5 for bringing out the audio lead. After making the electrical connections, some speaker cement

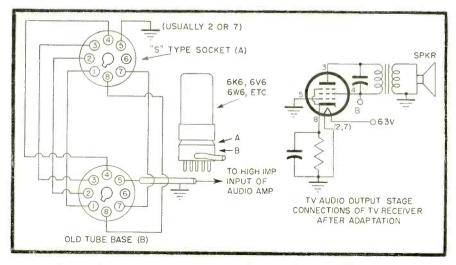


Fig. 1—Recommended when amplifier is close to TV receiver.

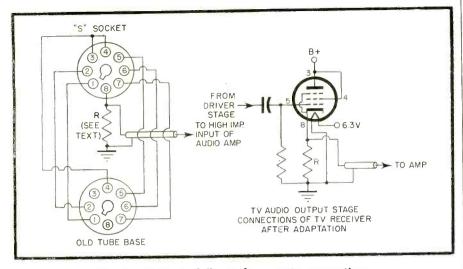


Fig. 2—Cathode follower for remote connections.



An "S" type Octal Socket.

should be applied where the socket and base meet. When the joint has dried, the original tube can be replugged into the adapter, and the adapter plugged into the power output stage of the TV set. It is important that the audio output then be connected to the high imped-

ance input of the audio amplifier. Overloading and distortion will result if the pre-amplifier input is used.

A metal tube of the same type can be adapted by the above method if precautions are taken to see that pin I on the tube socket has not been used as a dc tie-point by the manufacturer. This condition would render the tube shell hot to ground and create a shock hazard.

Adapter for More Remote Locations

Figure 2 illustrates a variation of the above which can be used when the TV receiver is remotely located from the audio amplifier. Before trying this one, however, the customer should be warned that having the sound come from a different point in the room from the picture might seem peculiar at first. Adjusting to it isn't difficult, though, and the improved quality will be worth the effort. As above, only octal tubes

[Continued on page 49]

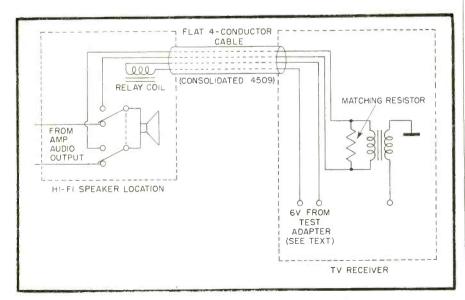


Fig. 3—Automatic switching of Hi-Fi speaker to TV output.



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2850 Irving Park Road, Chicago 18, III. "America's Most Complete Wire Line" THIS installment is devoted to three sync amplifier problems. The use of the oscilloscope is extremely important in hunting down these faults. Equally important is the manufacturer's service data which usually contains the correct waveforms at the important places.

Admiral 21C1—Lack of horizontal and vertical hold

The receiver was turned on and it was observed that the horizontal and vertical hold would break as the contrast was lowered; however, at maximum contrast, the picture would hold properly. As this seemed to be a composite sync problem, V403, 12AU7, the sync separator and clipper, and V401B, 1/2 6SN7, sync inverter were individually replaced but without effect.

Before going any further, the diagram was consulted. The composite sync, video and blanking voltage is fed through a coupling network consisting of R323, R329, C315, C308, to V403A. The low plate voltage (32 volts) on V403A, Sync Separator, developed by R418, and R417, and the cut-off grid leak bias provided by R414 and C308 allow only the most positive peaks (sync pulses) to cause plate current flow. In this manner the blanking and video voltage is eliminated and only the sync pulses are amplified. V403B is the sync clipper, and because of the low plate voltage caused by R420 and the grid leak bias caused by C409, and R416, the negative sync and noise pulses drive the tube to cutoff; thus achieving the desired clipping

The Work Bench

by PAUL GOLDBERG

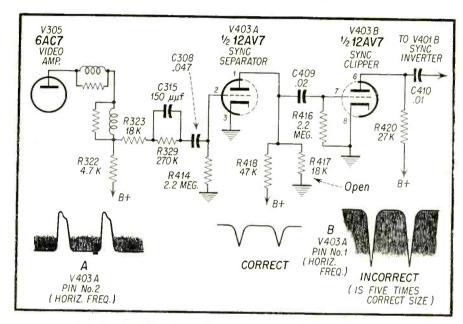


Fig. I—Partial schematic of Admiral 21C1 sync section.

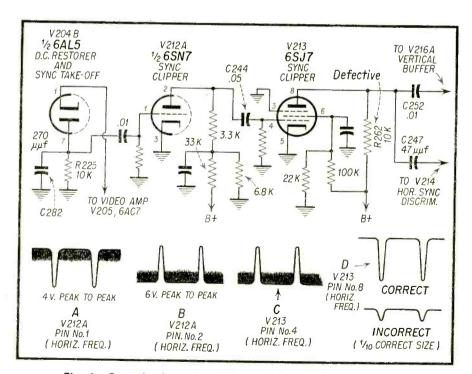


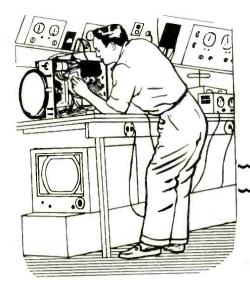
Fig. 2—Partial schematic of DuMont RA 103C sync section.

and limiting of the noise and sync pulses.

To get back now, the scope was set up to check a few wave forms. A wave form check was made at pin \$2 of \$403A\$, but it was found to check correctly with the manufacturer's service data. (Ref. to Fig. 1A). Next, a waveform check was made at pin \$1 of \$V403A\$. However, here the waveform did not check correctly with the manufacturer's data. (Ref. to Fig. 1B). In fact, the waveform measured about five times its correct size.

Now that the locality of the trouble had been placed, a voltage check was made at pin \$1 of V403A. The voltage measured 130 volts positive. The diagram called for 32 volts positive. Next, B417 and B418 were resistance checked. B418 measured correctly but B417, 18K, was found to be open. It was immediately replaced and the receiver now functioned properly.

Because R417 opened, the plate voltage jumped from 32 volts to 130 volts. With this plate voltage the negative grid leak bias developed was not



This Month:

SYNC AMPLIFIER PROBLEMS

enough to cut off V403A. Thus, the complete signal at the grid, pin \$2, of V403A, appeared at the plate, pin \$1. Therefore, because no elimination of the video and blanking pulses could take place in V403A, the video and blanking pulses triggered the horizontal and vertical oscillators, causing poor vertical and horizontal hold condition.

Du Mont RA 103C—Vertical roll and horizontal wobble

In this receiver it was observed that when the contrast was increased or decreased, even slightly, the picture would roll vertically. The picture also had a bad horizontal wobble. In fact, tapping any component in the horizontal sync or oscillator section caused the picture to wobble horizontally.

This was definitely a composite sync problem, so V212A, V213, and V204B,

the composite sync clippers and the dc restorer were replaced individually, but with no effect. The oscilloscope was next set up to make a few waveform checks and the diagram referred to.

Since negative composite sync pulses appear across R225, V204B, the diode rectifier, acts not only as a dc restorer, but as a sync separator. V212A, and V213 are the sync clippers. They manage to clip the signal at both ends. V212A, accomplishes this by virtue of a low plate voltage (+ 19 volts) and V213, 6SJ7, accomplishes this by virtue of a low screen voltage (+ 50 volts). In other words, the composite sync signal at both control grids drives the tubes into saturation and into cut off. Thus, the composite sync level is kept rather constant at the output of V213 because of the limiting action of these

A waveform check was made first,

V120A 1/2 6SN7 V118 V114B 6SK7 1st. SYNC AMP. 6SH7 1/2 6AL5 2nd. SYNC AMP. 270 µµf RESTORER SEPARATOR R156 4.7 MEG. R 228 6.8 MEG. +135 V. R153 1 MEG. 1 MEG Open 18 V. +135 V. +275 V. -18 V. -2 V. V119 PIN No. 8 (HORIZ. C V119, PIN No.4 CORRECT FREQ.) V 118 PIN No. 4 (HORIZ, FREQ.) PIN No. 8 (VERT (VERT. FREQ.) SHOULD BE 2 TIMES THE SIZE OF FIG. 3C INCORRECT FREQ.) 7 TIMES THE SIZE OF FIG. 3A

Fig. 3—Partial schematic of RCA 630TS sync section.

at pin \$\pm\$1 of V212A. Here the waveform checked correctly with the service data (refer to Fig. 2A). Next, a waveform check was made at pin \$\pm\$2 of V212A and here again the waveform was correct. (Ref. to Fig. 2B). A waveform check was now made at pin \$\pm\$4, the grid of V213. This was also correct (Ref. to Fig. 2C). At pin \$\pm\$8 the plate of V213, however, the waveform was found to be one tenth its proper size (Ref. to Fig. 2D).

Voltage measurements were next taken at the plate and screen of V213. The screen voltage checked correctly at around 50 volts positive, but the plate voltage measured about 300 volts positive. It seemed that R262, 10K, was not dropping any voltage because the B+ side of this resistor also measured about 300 volts positive. R262 was then resistance checked and was found to measure about 200 ohms. C244, C252, and C247 were next measured for voltage leakage because of the possibility of their causing the defective R262. But all these condensers checked satisfactorily. The 6SJ7, V213 also checked okay. Thus, it was assumed that the low wattage rating of this resistor was the cause of its failure. This half watt resistor was now replaced with a new 2 watt 10K resistor (to be safe), and now the receiver functioned properly.

It is easy to understand that if R262, 10K, the load resistor across which is developed the composite sync pulse voltage, drops in resistance to 200 ohms, the load voltage will not be large enough to cause the vertical and horizontal oscillators to be triggered properly.

RCA 630 TS—Pulling at top of pix and vertical slipping

The receiver was turned on and it was observed that there was a quivering and pulling at the top of the picture. It

[Continued on page 49]

THE DETROIT SERVICEMEN'S SHOW



Photos courtesy of Michigan Press Photo Service

A number of distributors get together to put on a show for servicemen primarily with unparalleled results.

THE First Detroit Electronic Parts Show, staged at the Whitehall Bldg., State Fairgrounds, Detroit, Michigan on November 2nd and 3rd, 1954, attracted 6100 Radio and TV Dealers and Servicemen from the Greater Detroit area. The show was a great success.

The twelve independent electronic parts jobbers in Greater Detroit who sponsored the show in co-operation with manufacturers and members of the Wolverine Chapter of Representatives feel highly satisfied to have been of service to the Radio TV Service-Industry.

The show was opened by Lt. Governor Clarence A. Reid and Peggy Lee, well known recording artist. Mr. H. Hirt, Secr.-Treasury of Radio Specialties Co. was chairman of the event. Governor Williams sent his best wishes by telegram.

Following are some of the comments from a cross-section of the industry:

(1) H. H. Rainier, Mgr. Electronic Product Sales, Sylvania Electric Products, Inc.

Sylvania wishes to congratulate the Greater Detroit Distributors on their first undertaking of a show directed to dealers. This kind of effort demonstrates genuine sincerity on the part of the electronic distributors to bring the servicemen into direct contact with manufacturers and their products. Sylvania as a manufacturer who has always felt a close alliance with the service industry, feels that this initial venture by the Detroit Distributors was an outstanding success. We at Sylvania have attended many similar shows and we feel the Detroit Show is one of the best we have attended. Best wishes for continued success.

(2) H. Kalker, General Sales Manager, Sprague Products Co.

Let me congratulate you and your

committee for a very fine job in presenting the first Dealer Servicemen Show in Detroit, sponsored by the Distributors and the Representatives. Despite the inclement weather, I was pleasantly surprised at the thousands of Dealers, Shop Owners and Servicemen who came to the show, their hungry desire for information was well proven by the fact that by late evening most Manufacturers ran out of thousands of their pieces of literature. It is my hope that you will continue this program on an annual basis and that you have the Distributors, with the help of the Manufacturer, man the booths as these people are really your customers and not direct customers of the Manufacturers.

I am sure that in the next few months, you will have found that this Show has helped considerably in cementing a more personal and friendly relationship in addition to a business relationship between the Distributors, Representatives and Manufacturers. I was very happy to have had the privilege and opportunity to see the show in person.

(3) John T. Thompson, Sales Mgr., General Electric Co.

You and your committee are certainly



Mr. H. Hirt, chairman of the show.

to be congratulated on the complete success of the first dealer parts show conducted in the Detroit area. As far as I know, this is the first cooperative show of this type ever held for dealers and their technicians in the electronics service industry. Its complete success reflects the hunger of the dealer and his television and radio service technicians for more product information and guidance from his distribution suppliers and their manufacturing sources. We in the Tube Dept. of the General Electric Co., felt it a real privilege to participate with you in this progressive industry step and wish to assure you of our complete support in future activities of this type.

(4) E. D. Powers, Exec. Vice President of Thordarson-Meissner

I am truly amazed at the tremendous outpouring of TV & Radio Technicians who attended the 2 day show. I have been to components expositions all over the United States and have never seen the equal of the attendance at the State Fairgrounds. This is especially phenomenal when you consider that this is the first show exclusively for servicemen and technicians of its type ever attempted in the entire country. It is indeed a tribute to the imagination, planning and splendid organization of the electronic parts distributors of Greater Detroit, under the very capable chairmanship of H. Hirt of Radio Specialties Co., 456 Charlotte. The Radio & TV parts distributors of the U.S. can well take example from this excellent and inspired type of good merchandising in bringing the manufacturers of complex equipment in direct contact with the ultimate users of their products. It cannot help but generate good will, trust and greater cooperation among maker, distributor and servicemen.

Radio-TV Service Dealer Video Speed Servicing Systems © Data Sheets

Mfr: Majestic

Model No. 121B

Card No. MAJ 121-1

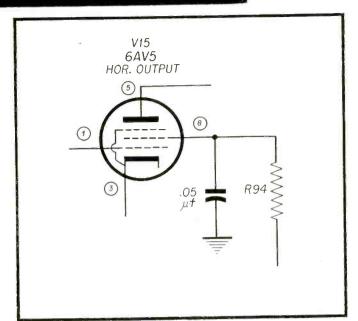
Section Affected: Raster

Symptom: No high voltage

Cause: Defective component

What to Do:

Replace: R94 (5K. 5 watt)-open



Mfr: Majestic

Model No. 121B

Card No. MAJ 121-2

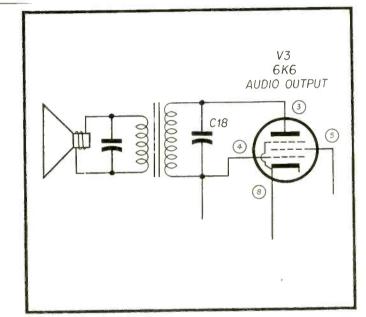
Section Affected: Sound

Symptom: No sound

Cause: Defective component

What to Do:

Replace: C18 (.005 µf)—shorted



Mfr: Majestic

Model No. 121B

Card No. MAJ 121-3

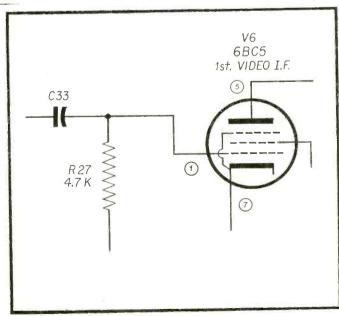
Section Affected; Pix

Symptom: Video overload

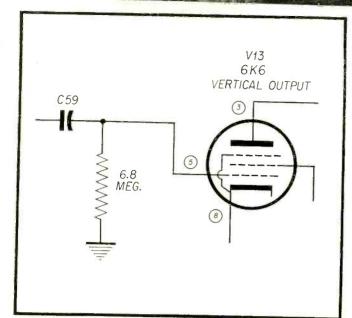
Cause: Defective component

What to Do:

Replace: C33 (100 $\mu\mu f$)—leaking



Radio-TV Service Dealer Video Speed Servicing Systems © Data Sheets



Mfr: Majestic

Model No. 121B

Card No. MAJ 121-4

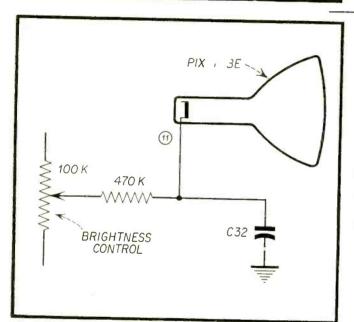
Section Affected: Raster

Symptom: Vertical foldover at bottom

Cause: Defective component

What to Do:

Replace: C59 (0.1 µf)—leaky



Mfr: Majestic

Model No. 121B

Card No. MAJ 121-5

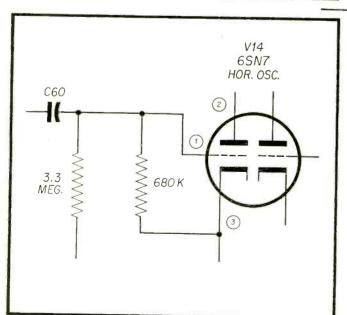
Section Affected: Raster

Symptom: No control of brightness

Cause: Defective component

What to Do:

Replace: $C32 (.05 \mu f)$ —shorted



Mfr: Majestic

Model No. 121B

Card No. MAJ 121-6

Section Affected: Sync

Symptom: Horizontal frequency drift

Cause: Defective component

What to Do:

Replace: C60 (.002 μf)—leaky

Radio-TV Service Dealer Video Speed Servicing Systems Data Sheets

Mfr: Philco

Model No. 50-T1600, 51-T1600, etc.

Card No: PH1600-1

Code No. 121 and 122

Section Affected: Sync

Symptom: Critical sync action under difficult

reception conditions

Reason for Change: Circuit improvement

What to Do:

Remove: 6.8K, 1.2 meg resistors as shown

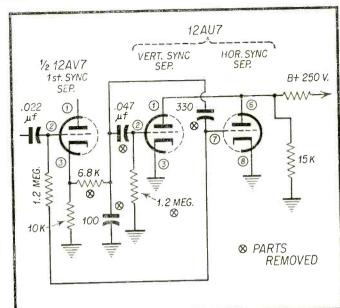
also $100~\mu\mu f$, $.047~\mu f$ and $330~\mu\mu f$ con-

densers as shown

Install: Parts and rewire as shown in Card

No. PH1600-2

(continued on Card No. PH 1600-2)



(continued from Card No. Ph 1600-1)

Mfr: Philco

Model No. 50-T1600,

51-T1<mark>600</mark>, etc.

Card No: PH1600-2

Code No. 121 and 122

Section Affected: Sync

Symptom: Critical sync under difficult reception

conditions

Reason for Change: Circuit improvement

What to Do:

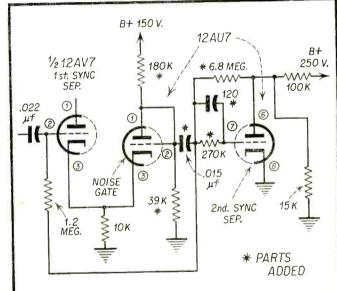
Install: 39K, 180K, 270K, 6.8K, 100K, resistors

as shown

Also, .015 μ f, 120 μ μ f condensers as

shown

Rewire: As shown



Mfr: Philco

Model No. 50-T1600,

51-T1600, etc.

Card No: PH1600-3

Code No. 121 and 122

Section Affected: Raster

Symptom: No raster, no high voltage; R104

(4.7K) resistor burns

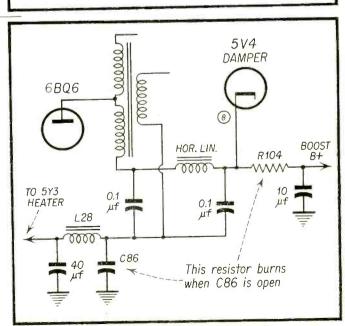
Cause: C86 (40 μ f) is open, causing horizontal rf energy to flow completely through

R104 (4.7K) which burns up.

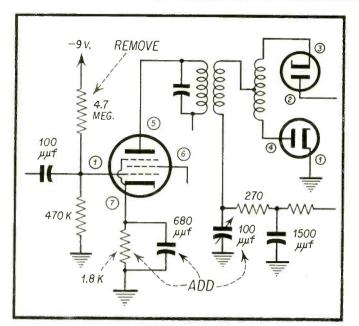
What to Do:

Replace: C86 (40 μf)

Also R104 (4.7K)



Radio-TV Service Dealer Video Speed Servicing Systems @ Data Sheets



Mfr: Philco

Model No. 50-T1600, 51-T1600, etc.

Card No: 50T-1600-4

Code No. 121 and 122

Section Affected: Audio

Symptom: Intercarrier buzz

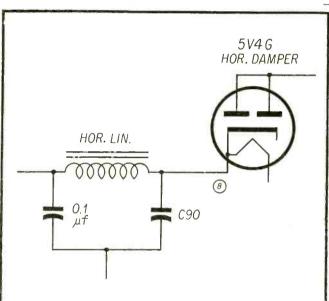
Cause: Insufficient AM rejection

What to Do:

Remove: 4.7 meg resistor Add: 680 µµf condenser

Also, 100 $\mu\mu$ f adjustable condenser and

1.8K resistor



Mfr: Philco

Model No. 50-T1600,

51-T1600, etc.

Card No: PH50T1600-5 Code No. 121 and 122

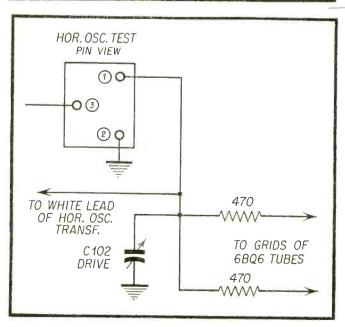
Section Affected: Pix

Symptom: Poor Horizontal linearity

Cause: Open condenser

What to Do:

Replace: C90 (.1 μf)



Mfr: Philco

Model No. 50-T1600,

51-T1600, etc.

Card No: PH1600-6

Section Affected: Raster

Symptom: Intermittent raster, bias on 6BQ6GT

intermittent

Cause: Defective drive capacitor (mica cracked)

What to Do:

Check: C102 (horizontal drive control), note

condition of mica and replace

Radio-TV Service Dealer Video Speed Servicing Systems® Data Sheets

Mfr: RCA

Chassis No. KCS 81

Card No: RC81-1

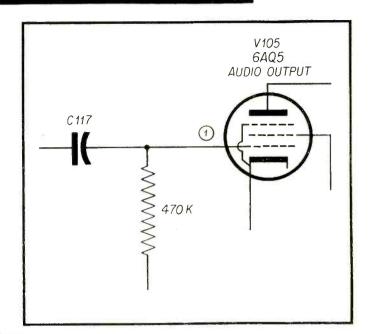
Section Affected: Sound

Symptom: Raspy sound

Cause: Leaky Condenser

What to Do:

Replace: C117 (.01 µf)



Mfr: RCA

Chassis No. KCS 81

Card No: RC81-2

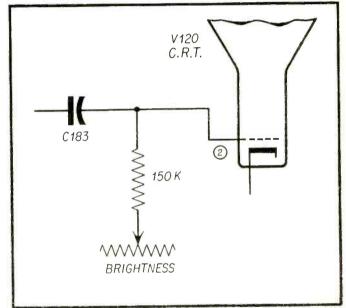
Section Affected: Raster

Symptom: No control of brightness

Cause: Shorted condenser

What to Do:

Replace: C183 (.1 μf)



Mfr: RCA

Chassis No. KCS 81

Card No: RC81-3

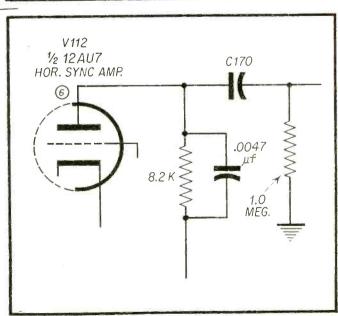
Section Affected: Sync

Symptom: Critical horizontal and vertical sync

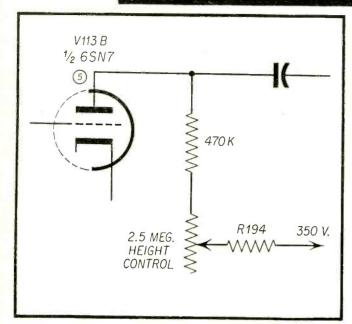
Cause: Shorted condenser

What to Do:

Replace: C170 (.01 µf)



Radio-TV Service Dealer Video Speed Servicing Systems Data Sheets



Mfr: RCA

Chassis No. KCS 81

Card No: RC81-4

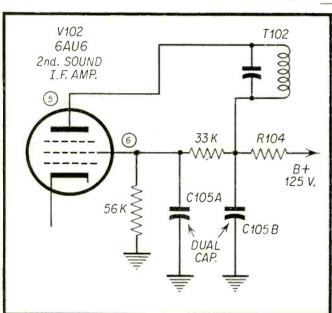
Section Affected: Raster

Symptom: Insufficient height

Cause: Resistor increases in value

What to Do:

Replace: R194 (220K)



Mfr: RCA

Chassis No. KCS 81

Card No: RC81-5

Section Affected: Audio

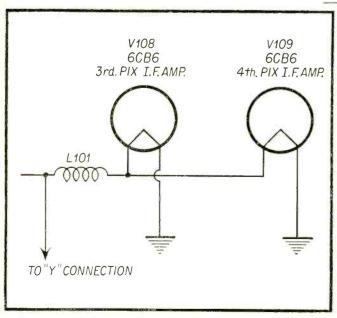
Symptom: No Audio

Cause: Shorted condenser; and resistor changed

in value caused by shorted condenser

What to Do:

Replace: C105B (.01 μf)
Also, R104 (IK)



Mfr: RCA

Card No: RC81-6

Section Affected: Sound & pix; V108, V109

Symptom: No sound, no pix

Cause: Open filament choke

What to Do:

Replace: L101 (filament choke)

PHICO

L301 (BOT) TC301 (Ø) J400

17

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VOLUME CONTROL SOCKET

6/

240

8

1C400

TC300

Code 140, Chassis R-191, D-191 TUBE LIST

R-191	FUNCTION
CHASSIS R-	CIRCUIT
RF CH	TYPE
~	SYMBOL

SYMBOL TYPE CIRCUIT FUNCTION	R-F Amp	Oscillator-	Mixer	Video I-F Amp	Video Amp	Video Output	Amp	Sound I-F	Amp	Ratio Detec-
TYPE CIR	6BZ7	8X9		75 6CB6	6.106	6.405		9.1V9		6.[3
SYMBOL	Vl	V_2		V3, V4, V5 6CB6	9/	$L\Lambda$		V.3		6/
C10	ь і	crr		~ г	D.F.	A 1 5	D		-	

Sound 1-1	Amp Ratio Detec-	tor, First Audio, and	Tuner A-G-C	Clamp	Audio Output	Sync Separa-	tor	17YP4 or Picture Tube	
9.71.9	6T8				99A9	9S29		17YP4 or	21ZP4A
00	6/				V10	VII		V19	

DEFLECTION CHASSIS D-191 SYMBOL TYPE CIRCUIT FUNCTION V12 12AU7 Phase Splitter, Vert. SYMBOL TYPE V12 12AU7

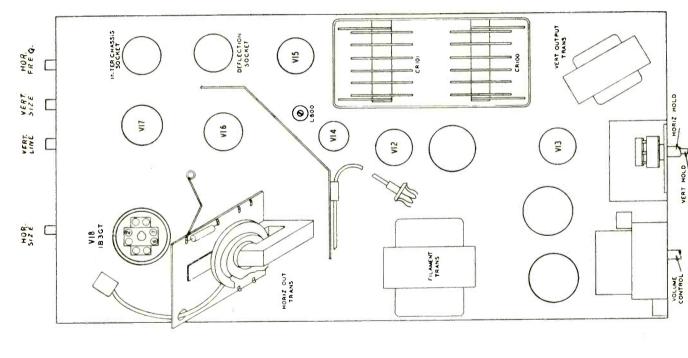
Uscillator	Vert. Output	Hor. Phase Com-	
	12B4	6AL5	
	V13	V14.	

parer	Hor. Osc.	Hor. Outp.
	12AU7	6B06GT
	V15	716

6AX4GT Hor. Damper 1B3GT H. V. Rectifier V18

KEY VOLTAGES

M con assis.	170 vd -38 vd	165 vd		205 vd	202 vd		420 vd		260 vd	
All voltages are measured with a VTVM connected between the tube pins and chassis.	pin 6 170 vd Grid of Hor. Out., V16 pin 5 -38 vd				149 to					
l with pins	pin 6 pin 5	pin 1		pin 9	pin 6		pin 3	per,	V17 pin 5	
easure e tube	, V16	V15	, c.,	, V13	V12	OSC.,	V17	of dam	V17	
are m	r. Out.		Hor. Os	rt. Out.		VERT.		cath.		
oltages d betw	of Ho		(s) of	of Ve		(s) of		ed B+		
All v	Grid		Plate	Plate		Plate		Boost		
42										



SPEAKER

AUDIO OUTPUT

45

VIDEO TEST

0

44

0

05

1/6

CHANNEL SELECTOR CHASSIS R-191, TOP VIEW, SHOWING LOCATIONS OF ADJUSTMENTS BRIGHTNESS CONTRAST R-FI

FINE TUNING

0

0

0

6

0

>

72

0

Deflection Chassis D-191, Base Layout

B SUPPLY FUSE REPLACEMENT

The B supply protective fuse, F100, is wired into the low-voltage section, and is in series with the selenium rectifiers. For replacement, use a 1.6-ampere delayed-action.

type fuse, Philos Part No. 45-2656-23.
CAUTION: Discharge the circuit before replacing the fuse.

HORIZONTAL-OSCILLATOR ADJUSTMENT

To adjust the horizontal-oscillator circuit, tune in a station and proceed as follows:

1. Reduce the width of the picture until approximately 1 inch of blank screen appears the right-hand and left-hand sides of the 2. Increase the BRIGHTNESS control setting until the blanking becomes visible. This appear as a dark vertical bar on each

side of the picture. S. Connect a. I. Mf. condenser from the test boint, adjacent to TC800, to ground. (The plate side of the horizontal ringing coil, L800,

is connected to the test point.)
4. Set the HORIZONTAL HOLD control to the approximate center of its mechanical ro5. Adjust the HORIZ HOLD CENTERING control unit equal portions of the blanking bar appear on both sides of the picture.

6. Remove the .1-\mu f. condenser from the test point.

Adjust the horizontal ringing coil, L800, until equal portions of the blanking bar again

out of sync on both sides of the center of its rotation. If the picture does not fall out of sync on both sides, readjust the HORIZ HOLD appear on both sides of the picture.

8. Rotate the HORIZONTAL HOLD conthrough its range. The picture should fall CENTERING control. trol

diagonal blanking bars that appear just before the picture pulls into sync. The pull-in should occur with from 1 to 2 diagonal bars when the sync position is approached from either direction. If proper pull-in is not ob-9. Rotate the HORIZONTAL HOLD control through its range, and observe the number of tained, repeat the above procedure,

VIDEO PEAKING-COIL ADJUSTMENT

or excessive smear is present, a slight adjustment of L303 may improve the picture quality The video peaking coil, L303, is adjusted at man. On any station where excessive overshoot factory for proper transient response of the video circuits. Ordinarily, this coil will on that station; however, this adjustment may L.303 is replaced in servicing, adjustment will require no further adjustment by the servicesacrifice the quality on other channels.

alignment and i-f alignment. (Never adjust be required. Before adjusting L303, check the

L303 until there are no trailing whites or smear in the picture. Turning TC301 clock-wise reduces trailing whites and overshoot; until the alignment of the receiver is correct.) Then tune in a station and adjust turning TC301 counterclockwise reduces picure smear and increases trailing whites. The proper position is the point where no smear or trailing whites appear in the picture.

The above procedure for adjustment of TC301 applies to a particular station exhibiting smear or overshoot. After TC301 is adjusted, reception on all the other stations should be checked, to make certain that the adjustment has not impaired the picture

OSCILLATOR ALIGNMENT

Tuning cores are provided in the oscillator coils at channels 13, 11, 9, 7, 6, and 4, By adjusting these tuning cores, all channels may be placed on frequency. This procedure should be carried out with the highest frequency channel first, since the alignment of each channel affects the alignment of all the channels below it in frequency. The channel adjustments are so arranged that, with one exception, each adjustment corrects the tuning of more than coverage of the various one channel. The coverage

CHANNELS CORRECTED BY ADJUSTMENT **ADJUSTMENT** CHANNEL

Contrast con.
V3, V4, V5, V6, V7
Check Vid. Det. and Amp. peaking coils
Check Vid. Det. xstal CD200 (1N64)
IF and RF alignment.

Check adjustment of TC301

13 and 12	11 and 10	9 and 8	7 only	6 and 5	0 0 0 0
13	11	·c.	ţ.	9	

all adjustments by placing the stop on the FINE TUNING cam between the Channel 7 and 8 holes on the front plate of the tuner. The FINE TUNING cam should be preset for

PROCEDURE USING STATION SIGNAL

Check adjustment of TC300

IF and RF alignment

Tuner fine tuning V1, V2, V3, V4, V5, V6

SOUND BARS IN PIX

The following simplified procedure may be used to align the oscillator when the television i-f alignment is satisfactory and a sta-1. Mechanically preset the FINE TUNING tion signal is available:

cam to the center of its range (see figure 4). 2. Tune in the highest-frequency channel to be received.

3. Adjust the tuning core for that channel, or the next higher channel, for the best picture; that is, starting with sound in the picture, turn the tuning core until the sound disappears. Repeat for each channel received in the area.

Power input circuit Check B+ Fuse F100 (1.6 Amps.) Check selenium rectifiers CR100 and CR101

NO RASTER-NO SOUND

PHILCO TROUBLE SHOOTING CHART

PIX BENDING
Hor. Hold and Fred. con.
University of TC800
VII. VI2. VI4. VI6. VI6
Check 2000 μμί and 0.01 μί caps. connected to pin 7 of V14

NSUFFICIENT BRIGHTNESS Ion trap

Contrast con. v3, V4, V5, V6, V7 Check Vid. Det. xstal CD200 (1N64)

DOOR HOR. LIN. Width con. V16, V17

WEAK PIX-SOUND AND RASTER OK

Tuner fine tuning

Brightness con. V16, V17, V18, V19 Low line voltage

EXCESSIVE RASTER (PIX SIZE)

Check 0.047 µf cap, connected to terminal

TI of hor, out, trans. Hor. Out trans.

PIX JITTER SIDEWAYS

Hor. and Vert. con. V16, V17, V18, V19

INSUFFICIENT RASTER WIDTH Width con.

Hor. Hold and Frey. con. Check adjustment of TC800 Check 0.01 μf cap. connected to pin 7 of

Check 1000 µµf cap, connected to pin 7 of

Check selenium rectifiers CR100 and CR101 Check 390 and 5000 $\mu\mu f$ caps, connected to pin 1 of V15 V16, V17

Low line voltage Hor. Out. trans.

INSUFFICIENT RASTER HEIGHT Vert. Size and Lin. con. V12, V13

Check Vid. Det. xstal CD200 (1N64)

Tuner fine tuning

SMEARED PIX

Vert. Hold and Contrast con-

PIX JITTER UP & DOWN

Check 0.1 µf cap, connected to pin 2 of V13 Check B+ voltage
Check 0.015 \(\mu \) f cap. connected to red lead
of Vert. Osc. trans.

Low line voltage

NO VERT. DEFL. V12, V13

Check 0.1 µf cap, connected to pin 2 of V13 Check 0.015 µf cap. connected to red lead Vert. Defl. coils (yoke) V. O. T. and Vert. Osc. trans. of Vert. osc. trans.

NO HOR. OR VERT. SYNC.—PIX SIGNAL OK

V3, V4, V5 Check Vid. Det. and Amp. peaking coils

Tuner fine tuning Focus con.

POOR PIX DETAIL

IF and RF alignment

Check 0.017 $\mu \mathrm{f}$ cap, connected to pin 2 of V12

NO HOR. SYNC.-VERT, SYNC.

š

Hor. Hold and Freq. con. Check adjustment of TC800 V12, V14, V15, V16 Check 390 $\mu\mu$ f cap. connected to pin 2 of

DISTORTED SOUND

Check 0.0068 µf cap, connected to pin 5 of Sound and Vid. IF alignment TC400, Det. Tuner fine tuning V2, V8, V9, V10 Check Vid. Det. xstal CD200 (1N64) alignment Z400

VERT. BARS VIG, VIT Cheek 68 unt cap, connected to yoke ter-minals XI and Y2

Check Vid. Det. and Amp. peaking coils

Check adjustment of TC301 Check Vid. Det. xstal CD200 (1N64)

Contrast con. V1, V2, V3, V4, V5, V6, V7, V19

ENGRAVED EFFECT IN PIX

Tuner fine tuning

STEWART WARNER

Model 9132-A

TUBE LIST

CIRCUIT FUNCTION 1st Sound IF Amp. SYMBOL TYPE

Sound Discr.-Sound 2nd Sound IF Amp. 6AU6 8L9 V3 $\sqrt{2}$

Amp. Sound Out. Mixer-Osc. TD9U9 9[9 75

lst Vid. IF Amp. RF Amp. 6B07 or **6BZ7** 6AU6 9NV9

2nd Vid. IF Amp. 3rd Vid. IF Amp. 4th Vid. IF Amp. 6AU6 6CB6 6AL5 V10V11 6/

Sated Sync. Sep. Vid. Det., DC Re-/id. Amp. storer 6AU6 6BE6

Picture tube AGC Amp. /id. Out. 6K6GT 21AP46AU6V14 V15 V16 V17

Hor. AFC Phase Det. Sync Slipper, Phase Splitter 12AT7 6AL5

Hor. Out. Hor. Osc. T97NS 99CD9 V18 V19 V20 V21 V23 V24 V25 V25

AV Rect. Damper 6W4GT 1B3GT

Rect.

6X5GT

Vert. Osc. Vert. Out. Rect. 6J5GT**5U4G**

PAING BON 1040-1v REAR CONTROLS PHONO

KEY VOLTAGES

V22 pin 5 330 vdc 480 vdc Boosted B+, cath. of damper, B+, plate of damper,

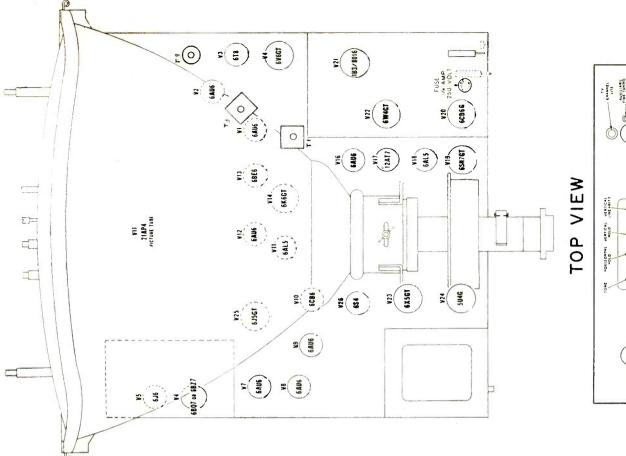
75 vdc V22 pin 3 V25 pin 3 Plate(s) of VERT. OSC.,

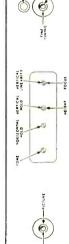
420 vdc Plate of Vert. Out., V26 pin 9

310 vdc 110 vdc Vl0 pin 2 pin 5 Plate(s) of Hor. Osc.,

V20 pin 5 -13 vdc Grid of Hor. Out.,

VTVM connected between the tube All voltages are measured with a oins and chassis.





FRONT CONTROLS

ADJUSTMENTS

ADJUST ION TRAP

If screen remains dark or is only dimly ilminated when "Brightness" control is turned clockwise, the ion trap may require adiustment.

The ion trap is located on the neck of the picture tube and consists of a magnet held in

of receiver) set approximately 1/2 turn clockwise, rotate the entire ion trap assembly while position by metal bands.
With "Brightness" control (located at rear sliding it back and forth until picture tube screen is illuminated to maximum brilliance. Reduce "Brightness" control setting and repent this operation to assure accurate positioning of iron trap.

maximum clockwise position until ion trap is correctly adjusted—failure to observe this precaution may result in damage to the pic-Do not turn 'Brightness" control to its

ture tube screen. If defocusing of picture or neck shadow results, DO NOT adjust position of ion trap. Failure to observe this precaution will result in an ion burn (brown spot) on the face of the picture tube. Since this condition is brought about by lack of proper ion trap CAUTION: There is only ONE correct setting of the ion trap. This position is attained when maximum brilliancy occurs on the picadjustment at time of receiver installation, the picture tube will not be eligible for warranty adjustment. ture tube.

AUXILIARY FINE TUNING ADJUSTMENT

It is found that the tuning range of the "Fine Tuning" control is inadequate to permit correct tuning of a station in its assigned channel, then adjustment of the "Auxiliary Fine Tuning" serew will be necessary. This special screw is accessible after removal of the "Channel Selector" and "Fine Tuning" knobs, They may be removed by merely pulling them forward.

screw may be undertaken in accordance with Adjustment of the "Auxiliary Fine Tuning"

the following procedure.

a. Set "Channel Selector" to desired channels:

then remove this knob. Set "Fine Tuning" knob to the center of portion of the main tuning shaft (outer shaft) should now be in the uppermost po-sition. Note the location of the "Auxiliary its range; then remove this knob. The flat Fine Tuning" adjustment screw on receiver

metallic), adjust the setting of "Auxiliary Fine Tuning" screw for correct tuning of the desired television station. CAUTION: Do not attempt to rotate this screw more than two full turns in either direction, as further rotation may release it from the bread clip within the tuning mechanism and the coil for that channel (located in R.F. Tuner Unit) would then have to be removed in order to restore the screw to Using a thin screwdriver (preferably nonchassis.

DECEMBER, 1954

pensated by resetting the "Television Fine range of the "Television Fine Tuning" control (after knob is replace of the shaft) will be adequate to tune in the shaft) will be adequate to tune in the the correct position. If a metal screwdriver is used, detuning occurs when the screw-driver is removed but it will be noted that this degree of detuning can now be comstation.

"Auxiliary Fine Tuning" screw for one channel. Identical screws are provided on each channel and they are all accessible the mechanism as each successively moves into position when the "Channel Selector" knob the same opening in the tuning Jo "Auxiliary Fine Tuning" screw for adjustment This completes the is rotated. thru

HORIZONTAL LINEARITY

the left edge of the screen and extended on the left edge or vice versa. Adjust for proper linarity by using "Horizontal Linearity" con-trol located at rear of chassis. In event that proper horizontal linearity cannot be obtained by adjusting this control, then check the setting of the "Horizontal Drive" control to be Improper horizontal linearity causes the circular test pattern to appear condensed on ting of the "Horizontal Drive" corsure it has been correctly adjusted.

HORIZONTAL DRIVE

a point just prior to where the picture tube brilliancy and size decrease. Then adjust this control ½ turn counter-clockwise. The "Horizontal Drive" control located at rear of chassis should be rotated clockwise to

ELIMINATING SEMI-CIRCULAR SHADOW

This shadow is caused by the electron stream striking the neck of the tube and it can generally be corrected by applying one or a combination of the following procedures:

Make sure deflection voke is symmetrically positioned with respect to the neck of the picture tube by loosening the three wing nuts and adjusting yoke position.

Should a change in height of the deflection oke be required to obtain symmetry of yoke sary to loosen the four wing nuts (two nuts located on each leg of mounting frame) and raise or lower voke as required. After adjustaround neck of picture tube, it will be necesnent has been completed be sure to retighten nuts securely.

BRIGHTNESS

The "Brightness" control (located at rear of receiver Chassis) should be initially adjusted in conjunction with the "Contrast" control for the brightness level desired by set owner check all active station channels. Too much little contrast, and vice versa, therefore, it is necessary to strike a proper balance between brightness will have the same effect as too ontrast and brilliance.

STEWART-WARNER TROUBLE SHOOTING CHART

FRT. BARS

Check 56 µµf cap, connected to yoke socket Defl. yoke ringing Hor. Drive con. terminal V20, V22

PIX BENDING

Hor. Hold. con. V16, V18, V19, V20 Check 0.05 μf and 5000 μμf caps. connected to pin 1 of V19 Hor. Hold. V16, V18, V

EXCESSIVE RASTER (PIX SIZE)

Hor. Drive con. Hor. and Size Sync. con. V15, V20, V21

RASTER BLOOMING

H Check HV Filter cap. Check 1 Meg Q Res. connected to Hor. Drive con. V15. V20. V21, V22, V23, V24 Filter cap.

INSUFFICIENT RASTER WIDTH

Hor. Drive and Size con. V20, V22, V23. V24 Check 270 and 390 $\mu\mu$ f caps. connected to Low line voltage Out. trans. pin 5 of V19 Hor, Out, trans

INSUFFICIENT RASTER HEIGHT Vert. Size and Lin. con.

Check 0.1 \mu f cap. connected to pin 6 of Check 0.05 μf cap, connected to pin 6 of V26 through a 6.8k Ω res. V25, V26 Out, trans. Low line voltage Vert.

NO VERT. DEFL. V25, V26

Check 0.1 μ f cap, connected to pin 6 of V26 Check 0.05 μ f cap, connected to pin 6 of Vert. Defl. coils (yoke) V. O. T. and Vert. Osc. trans. V26 through a 6.8k O res.

NO VERT, SYNC,-HOR, SYNC, OK

Check 4700 µµf cap, connected to pin 5 of Vert Hold con. Vert. Int. network V13, V16, V17

NO HOR. OR VERT. SYNC.—PIX SIGNAL OK V13, V16, V17

Check 0.05 µf cap. connected to pin 7 of V13 Check 0.05 µf cap. connected to pin 2 of V17 Check 330 $\mu\mu$ f cap, connected to pin 2 V19 NO HOR. SYNC.-VERT, SYNC. Hor. Hold con.

DISTORTED SOUND

Tuner fine tuning V1, V2, V3, V4, V6 Check 0.005 µf cap. connected to pin 5 of V1 Sound and Vid. IF alignment T1, T2 Det. alignment T3

Tuner fine tuning NO SOUND-PIX OK

Vol. con. V1, V2. V3, V4 Speaker (open voice coil or defective con-Sound and Vid. IF alignment T1, T2 Det. alignment nection)

Sound IF and Det. alignment T1, T2, T3 Tuner fine tuning V1. V2, V3, V11, V16

SYNC. BUZZ IN SOUND

NO RASTER-NO SOUND V23, V24

Check B+ fuse (1 Amp.) NO RASTER-SOUND OK

Power input circuit

V15, V19, V20, V21, V22 HV trans. Hor. yoke CRT connections Brightness con. Check HV Fuse (0.25 Amps.) lon trap

WEAK PIX-SOUND AND RASTER OK Tuner fine tuning

V5, V6, V7, V8, V9, V10, V11, V12, V14, V16 Contrast con.

POOR HOR. LIN.

Check 0.05 µf caps, connected to Hor. Lin. Hor. Lin. and Drive con. V20, V22 Hor. Out. trans.

POOR VERT, LIN.

Check 0.1 µf caps, connected to pin 6 of cap, connected to pin Vert. Size and Lin. con. V25, V26 Check 100 Mf Elec. Vert. Out. trans. 2 of V26 1.26

PIX JITTER SIDEWAYS

Check 0.01 and 0.05 µf caps. connected to Hor. Hold Con. V18, V19, V20 pin 1 of V18

Tuner fine tuning MEARED PIX

Check Vid. Det. and Amp. peaking coils Check 0.05 \(\text{mf} \) cap. connected to pin 5 of V14 Contrast con. V7, V8, V9, V10, V11, V12, V14, V16 IF and RF alignment

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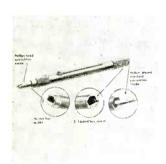
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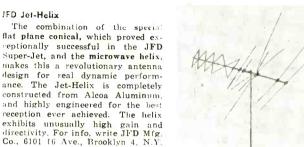
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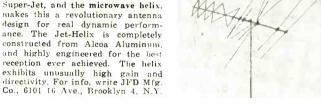
CBS-Hytron 4-Way Tool

CBS-Hytron, Danvers, Mass., has announced a 4-Way Tool consisting of: a slotted steel barrel with hex sockets, 1/4-inch at one end, 5/16-inch at the other; a double-ended screwdriver blade . . . one end with Phillips head, the other standard . . which slides inside the barrel: and a knurled setscrew to lock the and a kituried setserew to lock the sliding blade into any one of three recessed positions within the slotted barrel. When the sliding blade is centered within the barrel, either liex socket can be used.



Erie "Pallet-Pak"

Erie Resistor Corporation, Distributor Division, 50 E. Wynnewood Rd., Wynnewood, Pa., announces new method of bulk packaging ERIF. Disc Ceramicons at no extra cost, which it has named "Pallet-Pak." Five discs are inserted between two strips of cardboard and stapled in position between the leads, eliminating tangled leads and providing an easily identified, easily handled, and



quickly inventoried item.



Snyder Mike Stand

A new floor stand has been inroduced by Snyder Manufacturing Company of Philadelphia. The stand is a floor model and is coded the MS-2. 32½ inches collapsed, its chrome plated 2-section telescopic staffs extend to 61½ inches and are held securely by the Snyder Sure-Grip Lock with Velvet Action. The Base, of gray crackle finish cast iron, is light in weight. Additional information may be obtained by writing to Dick Morris at Snyder Mfg. Co., Philadelphia 40, Pa.



speaker driver unit.

Clear Beam Antenna Corp. announces availability of the "Big Chief," an all new high gain antenna combining the principles of the giant conical with a Yagi. This antenna provides servicemen with a functional type high gain all-band unit designed to cover all channels and special die-formed hardware provide easy assembly. The use of wider diameter tubing on the low band section of the antenna affords improved sensitivity



Raytheon Set Jackets

"Make your TV set match your room decor" is the theme of a new display featuring the "jacket wardrobe" of Raytheon's 17-inch and 21inch Challenger sets. The full-color display measures 2 by 3 feet, stands securely on top of the 17-inch set through a combination of die-cut design and easel back, and contains swatches of the fabrics used for the ten slip-on jackets. For details, write Raytheon Mfg. Co., 801 N. Michigan Ave., Chicago 11, Ill.



A new FM-AM professional tuner. Model AF-250, is now being manufactured by Regency which utilizes a genuine Armstrong circuit with two limiters, and has separate FM and AM circuits from antenna to and AM circuits from ancenna to output. AFC on FM provide maximum fidelity at all times. The unit is of cast polystyrene over black krinkle steel. For information, write Regency, a division of I.D.E.A., Inc. 7900 Pendleton Pike, Indianapolis



Cornell-Dubilier Flyback Checker

The CD "Flyback Transformer and Yoke Checker" is extremely sensitive and can detect even a single short circuited turn in the windings by indicating inductance change. The instrument, called the CD model BF-80, employs an oscillator circuit, a 6V6 vacuum tube and a 4½" microammeter with separate indi-cator scales for short tests, continuity tests, and yoke tests. For further information write: Cornell-Dubilier Electric Corporation, South Plainfield, N. J.



Development of the Roto-King new antenna rotator has been announced by Channel Master Corporation, Ellenville, N.Y. The rotator is designed to provide the set-owner with hair-line accuracy in pin-pointing TV stations. Its many features include a flexible worm gear which eliminates the danger of gear-slip-ping, back-lash, and binding, assuring a longer life and positive control, and its midget control cabinet, the smallest cabinet on the market, measuring just 2¾" x 4"





WORK BENCH

[from page 35]

was also noted that when the contrast was decreased slightly, the picture would slip vertically. Evidently, this was a composite sync problem. Therefore, V118, 6SK7, the first sync amplifier, V119, 6SH7, the sync separator, and V120A, the second sync clipper were replaced individually but did not solve the problem. Just as a slight possibility, V114B, 6AL5, the dc restorer was also replaced

The oscilloscope was next set up to check a few waveforms, and the diagram consulted. The 6SK7, V118, is a temote cut off type tube which is grid biased by the bleeder network, R228 and R153. The signal arriving at the grid of V118 is in a negative direction.

Knowing these facts, a waveform check was made at pin \$4 of V118 but the waveform checked correctly with the manufacturer's service data (Ref. to Fig. 3A). Next a waveform check was made at pin \$8 of V118 and this waveform also checked correctly with the manufacturer's notes (Ref. to Fig. 3B).

V119, the sync separator was then considered. This tube is grid biased by the bleeder network R156 and R157. The operating voltages applied to the grid, screen and plate are such that the negative or bottom portion of the applied positive going signal is cut off. Thus, the video and blanking pulses are removed and only the sync pulses appear at the plate of VII9. Knowing this, a waveform check was now made at pin \$4 of V119. This waveform checked correctly, vet it seemed to quiver a bit too much (Ref. to Fig. 3C). However, on checking pin #8, the plate of V119 we found trouble. Instead of being twice the size of that at pin #4 of V119 the waveform was about the same size. Moreover, it appeared to quiver entirely too much (Ref. to Fig. 3D).

A voltage measurement was then taken at the grid of V119, pin \$\frac{1}{24}\$ Instead of measuring about \$-5\$ volts, it measured about \$-18\$ volts. The receiver was then turned off and a resistance check was made of \$R156\$ and \$R157\$. \$R156\$ checked properly but \$R157\$ measured infinite. \$R157\$, I megohim, was then replaced with a new I meg resistor and the receiver now functions.

tioned properly.

With R157 open and —18 volts applied to the grid only the uppermost positive tips of the sync pulses would cause V119. 6SH7, to conduct. Thus, because of the small sync pulse voltage applied to the horizontal and vertical oscillators, the picture would slip vertically and pull horizontally.

AUDIO ADAPTATIONS

[from page 33]

such as the 6K6, 6V6, 6W6 etc. can be

adapted.

Again an "S" type socket and an old tube base are connected together, but this time as shown in Fig. 2. This results in the plate and screen of the output tube of the TV receiver being connected together so that the tube performs as a triode. Observe also that the

plate of the tube is connected to B+, and that the signal output is taken off the cathode. Thus, this stage becomes a cathode follower with a low impedance output which may be connected into a low impedance load. Systems between signal sources and loads using low impedance circuits produce a minimum of noise and interference voltage pickup in the lines, so that long lengths of wire may be connected between the cathode output terminals and the distant power amplifier input without excessive noise or hum pickup.

[Continued on page 52]





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Television CRT manufacturers' sales of cathode ray tubes in September topped the 1.1 million mark to establish a new high in monthly unit sales, as reported by the Radio-Electronics-Television Manufacturers Association. Receiving tube sales in September gained by over 16 percent from August and were at the highest level since June of last year. Nearly 1 million television receivers were sold at retail during September, bringing sales for the first nine months of this year to the highest point on record. TV set sales in September more than doubled the August rate and radio sales were at the highest level this year, the Association added.

The Sprague Electric Company has started construction of a 13,000 square foot, one-story building in the Venice section of oLs Angeles, Calif. to house all its Southern California operations. The building, in contemporary modern California architectural style of stone, glass, and cinder block, is epected to be completed in the spring of 1955. The Sprague company will then move from its present rented quarters in Culver City.

One-year warranties on TV picture tubes appear to be on the way out. On December 1, GE and DuMont will switch to six-month warranties on all sales of replacement CR tubes. A change to 12-month pro-rata warranty systems has already been made by RCA and Sylvania. The industry policy in regard to tubes sold to set manufacturers, but it is possible that these warranties may be reduced to a period of six months, early next year.

A new record care kit, ready in time for Christmas business, has been announced by Walco Products, Inc., East Orange, N. J. Designed to appeal to anyone with a record collector or hi-fi fan on his (or her) Christmas list, the kit contains a 6 oz. can of Walco Stati-Clean, popular anti-static spray cleaner for records; 2 packages of Walco Discovers—protective record sleeves made of the soft plastic, polyethylene; a book on record care by a leading audio expert, and a small camel's hair brush for cleaning dust from phonograph needle tips.

Latest developments in the Color TV field: Initial quantities of a pre-production run of RCA Victor's new 21-inch color television receiver, for demonstration purposes only, have been shipped to distributors in key color TV Markets. . . . Muntz TV, Inc., hopes to have a color set on the market in February, Earl W. Muntz, president, said recently. The color tube used will probably be CBS-Columbia's. . . . Magnavox will venture into color television for the first time next year with a 21-inch console incorporating a 21-inch round shadow mask color tube. . . . The U. S. Patent Office has granted Dr. Ernest O. Lawrence, a Nobel Prize physicist, a patent covering a basic invention for better and less expensive color television picture tubes. All Lawrence tubes embody the basic principle of post-deflection focusing and electronic color registration which makes it possible to build color stability in the tube itself. . . .

The "Dragnet" TV show, about the first of next year, will feature an exposé of a television service "racketeer." A. W. Bernsohn, managing director of the National Appliance and Radio-TV Dealers Association, announced that the script has been prepared by Dragnet Productions, Inc., with the cooperation of the Los Angeles Electric League and NARDA. Frank J. Moch, NATESA president, voiced apprehension at the announcement of the project, recalling that one of the earlier "Dragnet" scripts on the same subject proved quite derogatory to the industry; however, he expressed the hope that "anything controversial" may have been avoided this time by virtue of NARDA's editing of the script.

As of late, a great number of radio and television set manufacturers have been striving to build more compact units. As a result of this, the speaker must, of course, also be reduced. Since the dimension most important in this respect is the depth, there is an ever increasing number of shallow speakers being used. So as to offer the serviceman a suitable replacement speaker for these sets, Oxford electric has developed a series of shallow speakers which is identified as the "H" Series. These units have all the characteristics of the conventional replacement speakers. The "H" Series is limited to five sizes, which will be sufficient for almost all replacements.

A turn-out of some 250 service people attended the first 1954 meeting of the Service Managers' Organization of New York held at the New York Trade School on Wednesday, September 29th, according to Donald Packer, Service Manager of Olympic Radio & Television Inc., who sponsored the meeting.

OLYMPIC'S Chief Engineer, Lee Seigmund, spoke about service considerations of OLYMPIC'S new vertical chassis.



Norman B. Krim, Vice-President and General Manager Raytheon's Receiving and Cathode Ray Tube Operations, congratulates F. Edward Anderson, Distributor Tube Sales Manager, on completion of 25 years of continuous service. Mr. Anderson directs the replacement sales of Receiving, Picture and Industrial Tubes, including Semi-conductor Diodes and Transistors to over 600 distributor accounts. Mr. Anderson is a member of the Radio Oldtimers Club and is well known throughout the radio and television industry.

An agreement by GE field TV-radio service executives has been reached by which GE radio repair work will be channelled through independent service shops throughout Pennsylvania. A ninety-day free-labor policy has been inaugurated by GE for its radio line, by which all sets will be serviced by their distributors during the length of the guarantee period. GE, during a recent meeting with the Federation of Radio Service Men's Associations of Pennsylvania, declared that it did not want to exploit the service business and compete with service and appliance dealers.

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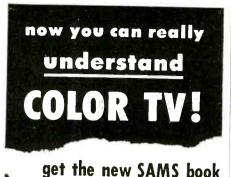
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AUDIO ADAPTATIONS

[from page 49]

The TV receiver schematic must be referred to for two reasons. The first is to make sure that the cathode resistor of the output stage is returned to ground -not B minus. Those circuits employing an ungrounded B minus supply do not lend themselves to this adaptation due to the high voltage difference between the two chassis. The second is to note the value of the cathode bias resistor. Since there is negligible dc drop in the TV audio output transformer, the B plus screen voltage can be applied to the plate without disturbing the rest of the circuit-thereby allowing the use of the cathode load resistor, R, of the same value as the old one.

The same materials are used as described in the previous section with the addition of the load resistor. Notice that the cable is connected to the socket this time, instead of the base. The cathode follower's inherent degeneration also makes it desirable from the point of view of reduced amplitude distortion. This increase in fidelity, however, is obtained at the expense of a loss in gain. With the rather low value of load resistor used here, the stage gain is somewhere between 0.35 and 0.50—low even for a cathode follower. This should not prove a serious drawback in most cases, however, since the audio amplifier has ample reserve gain to make up for this

Although most amplifiers contain an input grid coupling condenser, it pays to check the schematic for its presence. If there isn't any, a .01 uf paper condenser should be inserted at the amplifier end of the cable. Without this condenser, the grid would be driven several volts positive, causing excessive plate dissipation and distortion.

If, as is probable, these changes result in a duplication of volume controls, the amplifier control should be set at some high level and the volume varied at the TV set. This prevents the possibility of overdriving the earlier voltage in the receiver to make up for the loss in the cathode follower.

Speaker Switching by Means of Relays

If the customer's TV set uses a miniature audio output tube, such as a 6AQ5 (in which case the above adapters cannot be used because of the glass-to-metal seal) or his is one of those rare sets with a fairly good push-pull audio section, a great improvement in performance can be obtained by substituting the hi-fi speaker of his hi-fi system for the small commercial one used in most receivers. Fig. 3 indicates a method by which this

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can be done without the addition of any switches or controls. Here a relay automatically throws in the hi-fi speaker only when the TV set is turned on. A doublepole-double throw, 6 volt ac relay is mounted on the rear of the hi-fi speaker enclosure and connected as shown.

It is not advisable to mount the relay at the TV set since this might lengthen the speaker cable considerably, possibly causing high frequency oscillations in feedback type amplifiers. Four conductor flat cable is practical for this job since it can be laid under the carpet.

Voltage for energizing the relay coil is obtained from the filament terminals of any 6 volt tube in the receiver. These terminals are made accessible by the use of a Vector test adapter, type "T", which is plugged between the tube socket and the tube.

This will not work with a set with either series or series-parallel filament strings, since the filament current would be changed. It is also a good idea to choose a tube whose filaments are returned to ground rather than some positive dc voltage. High dc voltages can cause trouble if the relay frame is grounded to the chassis by its mounting stud.

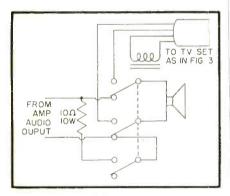
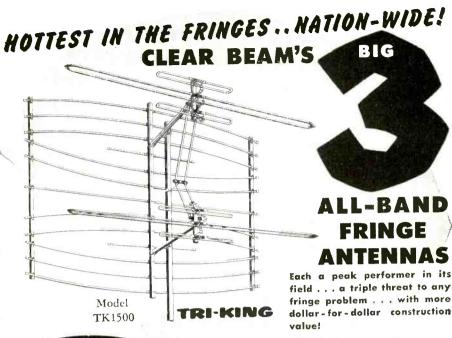


Fig. 4—Use of 3-pole relay to maintain permanent loading of audio output transformer.

Constant Loading on Output **Transformer**

With the above arrangement, it is possible to damage the hi-fi audio output transformer if the hi-fi audio system is turned on while the TV set is on. Fig. 4 shows how this hazard can be avoided at a slightly greater expense. By using a three-pole, double-throw relay, the transformer is loaded with a 10 ohm, 10 watt resistor when the TV set is in use, thereby maintaining a load on the hi-fi output stage under all switching conditions.

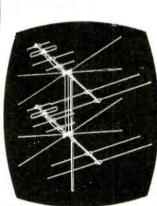
The matching resistor of Fig. 3 will be necessary when the speaker has a different impedance from the TV output transformer. A 10 ohm, 5 watt resistor is a good compromise for most cases.





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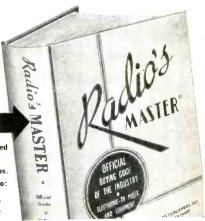
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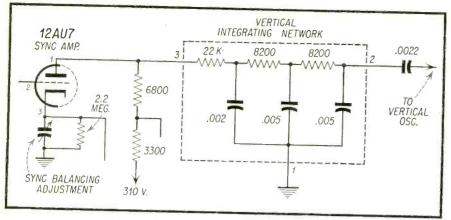


Fig. 2—Correct integrator wiring circuit.





The Miller K-Tran 1.F. Transformers are available for the following frequencies: 262 KC, 455 KC, 1500 KC, 4.5 MC and 10.7 MC.

4.5 MC transformers are for use in television receivers having an intercarrier sound channel. 10.7 MC transformers find their main application in FM receivers and tuners.

All transformers are shell core permeability tuned, thus providing a magnetic shielding of the windings and reducing the influence of the aluminum can. Stable silver mica fixed capacitors are enclosed in the low-loss terminal base.

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12-H2	262 KC	Output Transformer	1.50
12-H6	262 KC	^	1.50
		Output Transformer diode filter	1.59
12-C1	455 KC	Input Transformer	1.32
12-C2	455 KC	Output Transformer	
12-C6	455 KC	Output Transformer diode filter	1.32 1.41
12-C7	455 KC	Input Transformer for Battery Radios	
12-C8	455 KC	Output Transformer for Battery Radios	1.32
12-C9	455 KC	Input Transformer for Battery Radios	1.32
12-C10	455 KC	Input Transformer for AC-DC Radios	1.32
12 010	433 KC	Output Transformer for AC-DC Radios	1.32
13-W1	1500 KC	Input Transformer	1.44
13-W2	1500 KC	Output Transformer	1.44
13-PC1	455 KC	Input I.F. Transformer For Printed Circuit	
13-PC2	455 KC		s 1.44
	455 KC	Output I.F. Transformer For Printed Circuit	s 1.44
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ANSWERMAN

[from page 27]

of manufacturers TV receivers. But they do fail occasionally. What happens is that they open up inside at one of the leads. In the installation of the replacement integrating network several technicians have wired the new part in backwards.

Consider what would happen in Fig. 2 if the point 3 were connected to the .0022 uf condenser and point 2 were connected to the 6800 ohm resistor. The .005 uf condenser of the integrator network would short most of the horizontal pulse to ground leaving very little pulse to be used by the horizontal oscillator for lock-in purposes. However, the reactance of the .005 uf condenser to the vertical 60 cycle pulses would not be too low and there would be enough signal available at the output to hold the vertical oscillator in sync.

COLOR

[from page 14]

than that introduced by the filter in the I demodulator. To make both the I and Q signals arrive simultaneously at the matrix network a slight additional delay network is introduced in the I demodulator. Still on the subject of time delay it would be appropriate at this time to point out that the luminance channel which has negligible time delay is also provided with a time delay network. The latter brings the luminance signal into the matrix network at the exact same time as the I and Q signals.

Phase Splitter

In the I channel the I voltage is amplified and passed on to the phase splitter, whereas in the Q channel no amplification is necessary so that the output from the Q demodulator is coupled directly to its phase splitter. The purpose of the two phase splitters is to obtain both positive and negative polarity I and Q signal voltages. The positive and negative I and Q signals now enter a matrix network into which the Y signal is also fed.

Five signals are now available at the input of the matrix network. These are +I, -I, +Q, -Q, and Y. These five signals combine with each other in the matrix to produce the original red, green and blue color signals. The mathematical relationships are as follows:

R = .96I + .62Q + Y B = -1.101I + 1.7Q + YG = -0.281I - .64Q + Y

These primary color voltages are then fed to the color signal amplifiers which are controlled by potentiometers, except for the red amplifier stage which employs a fixed gain network. These potentiometers are set entirely by the requirements of the phosphor efficiency of the color picture tube and their adjustments will be discussed in detail in a later chapter.

Following each individual color amplifier a color output stage may be found depending on the receiver design. The red, blue and green voltages are then applied to the respective electron guns of

the tri-gun picture tube.

DC restoration is incorporated at this point. The use of dc restorers is essential to the proper presentation of color pictures. Whereas the omission of the dc component in a monochrome receiver is not serious because a turn of the "Brightness" knob can compensate for this loss, such a simple correction is not available to the operator of the color receiver. Although the "brightness or background" control of the color receiver does control all three guns simultancously it does not affect them equally. The ratio of the different gun effects are dictated by the relative phosphor efficiencies and cannot be used to correct for defective video presentation, as would occur in the absence of the three dc components.

REMOTE CONTROL

[from page 29]

- 6. Now place the unit back in the chassis and pick up location of mounting holes from the mounting bracket making sure that the shafts are perfectly aligned and the gap above-mentioned is maintained. Then drill two holes with a \$19 drill.
- Before making units secure to chassis, determine the proper method of coupling the tuner shaft (Figure 6).
- 8. Placing the Motor Assembly back in chassis, slide it forward so that coupler engages both shafts and secure brackets in position, #8 screw assemblies.
- 9. Where the method in Figure 4 applies, take the following steps: Place the base of mounting bracket A against rear wall of tuner so that large hole in the base centers around the tuner shaft. (NOTE: If the tuner shaft does not protrude through rear wall of tuner, it will be necessary to drill a hole in rear wall of the tuner so that the coupler can slide onto the tuner shaft without binding.)
- 10. Pick up mounting hole locations from mounting bracket and drill two \$29 holes in the tuner case. Mount with two \$8 sheet metal screws and lock washers.
- 11. Mount the Motor Assembly to

bracket B as shown in Figure 4 and secure with two \$8 screws and lock washers into the mounting studs of the Motor Assembly. (Motor Assembly in Figure 4 can be rotated 180° and then mounted if space so demands.)

- 12. Bracket B may now be mounted onto Bracket A dependent upon space limitations. The motor shalt should be cut at whatever length is desirable so that Bracket B can be bolted to Bracket A as space allows. It is important that a gap of from 1/32" to 1/8" be maintained between the shaft of the tuner and the shaft of the motor.
- 13. Now attach the coupler as shown in Figure 6A.
- 14. Slide the coupler onto the tuner shaft and bolt the brackets in place.
 Use \$8 screw assemblies.
- 15. Mount the Relay Assembly in any convenient location preferably on the rear of the chassis using #8 screws. (Bear in mind that the female disconnect plug will require a corresponding hole in the cabinet-back.)
- 16. The unit is now ready for wiring. Follow the wiring diagram, using tag attached to the Relay Assembly for sequence of wiring. PAY STRICT ATTENTION TO



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THE COLOR CODE. Be sure to recheck your wiring. Then plug REMOT-O-MATIC male plug into female plug on the Relay Assembly. On the remote unit, set the fine tuning knob up to its maximum (clockwise). Turn the set on at the set itself and align all operating channels to the center of the fine tuning range. Now shut the set off at the set and turn the brightness control on the set itself to its maximum position (clockwise).

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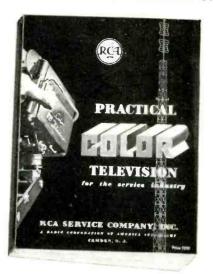
[from page 17]

is the subject of our discussion is termed, as previously mentioned, a "horizontal output and high voltage" transformer, or simply a "flyback" transformer.

Figure 4 is a simplified schematic for the purpose of explaining the production of the high voltage to be used for the 2nd anode of the picture tube. A logical starting point in this explanation is at the grid of the horizontal output tube. The driving voltage at this point is the modified sawtooth or trapezoidal wave of voltage shown in Fig. 4B. The horizontal output tube is biased with a voltage sufficiently negative so that current will flow through the tube only when the signal rises above the dotted line of Fig. 4B. The output tube therefore conducts between A and C, D and F, and G and I. It is cut off between B and C and also between D and E. It can be seen then, that the current in plate circuit will have the form shown in Fig. 4C. This is the current flowing through the section of the autotransformer winding from top #2 to top #1 in Fig. 4A.

As the plate current rises from A to B, a similar current is induced in the secondary circuit; similar that is, in wave form. Actually, the current in the secondary circuit is larger than that in the primary because of the step down ratio in the transformer. If this sounds strange, just recall that when the voltage is stepped down, the current is stepped up. As this current flows through the horizontal deflection coils, it builds up

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a steadily increasing magnetic field around these coils. This of course is the field which is responsible for the deflection of the electron beam. At points corresponding to B, E, and H of Fig. 4C, the magnetic field reaches its maximum value and then suddenly drops to zero. (Note. Actually, a shock excited oscillation begins at points B, E, and H. In the interest of simplicity and clarity, this action of the circuit will be omitted now and discussed in detail when the damping circuit is taken up.) Since the collapse of the field is very rapid, and since one of the factors which determines the size of the induced or back e.m.f. is the speed with which the lines of force cut the turns of wire in the yoke, there results across the yoke a high pulse of back e.m.f. This pulse reaches a value of approximately 1500 volts.

Notice, from Fig. 4A, that since this pulse appears across the terminals of the deflection coil, it must also appear across the terminals of the secondary winding of transformer, since these points are electrically the same. Now it is important to note that when the signal was being traced through from left to right, the transformer was a step down transformer, that is, fewer turns on the secondary than on the primary. However, in following the pulse of back e.m.f. from right to left, starting at the yoke, the transformer, working in reverse, steps up this rule of voltage. Terminals A and B, in Fig. 4A for example may be considered as the primary for the pulse, and terminals 1 and 3 the secondary. Depending on the design of the transformer, the pulse may be stepped up to a value of from approximately 9 KV for some of the 10 inch receivers to as high as 20 KV for some present day receivers.

[To Be Continued]

RESISTIVE NETWORKS

[from page 31]

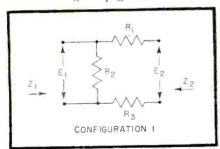


Fig. I—Unbalanced to balanced load network.

value of loss encountered can normally be compensated for by merely advancing the signal generator output.

An impedance match is not con-

sidered essential when a fixed-frequency generator is being used since the mismatch would normally only introduce errors in the indicated microvolt output, a measurement which is not accurate anyway with inexpensive signal generators. It follows of course that when accurate signal strength measurements are being made a proper generator-toreceiver impedance match should be provided. The loss introduced by the matching network must of course be taken into account in determining absolute values of receiver input.

This simple little device may take on considerable importance in uhf work. While tuner alignment has not constituted a major problem in vhf receivers it seems possible that it will in uhf equipment. If so, a matching device effective over a wide frequency range will be necessary for satisfactory results. It may also be necessary in recovering sufficient signal voltage to adjust antenna coupling in uhf tuners to provide proper antenna transmission line impedance matching, in which case it will be absolutely essential to have the test line properly terminated.

Transmission Line Termination

The antenna input impedance of a TV tuner may vary widely (as much as 10 to 1) for different channels. Hence

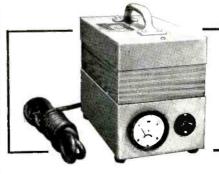
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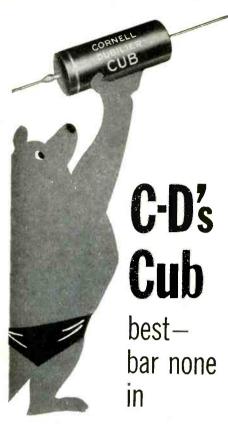
age Adjustor, either the inexpensive manual type or the deluxe automatic design.

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any signal generator-to-receiver impedance matching device which depends on the reflected impedance of the tuner input will be subject to considerable mismatch. For instance, with either a transformer match or a generator with a direct 300 ohms output a tuner input impedance of 3000 ohms would reflect a 10 to 1 mismatch to the transmission line. Under these conditions, alignment procedures could not be expected to give satisfactory results. It is therefore necessary to use an isolation termination such as Network C with direct output, transformer-matched or "balun"-matched sweep generators.

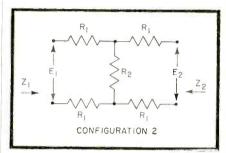
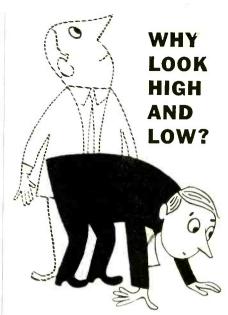


Fig. 2—Isolation network for regulating impedance across generator as load impedance varies.

With this isolating network designed for input and output impedances of 300 ohms the input impedance will vary only from 267 ohms to 370 ohms as the tuner input impedance varies from 30 to 3000 ohms. It is interesting to note that with Network A, designed for 72 ohms, a like change of tuner input impedance would cause the transmission line termination to vary only from 64 to 80 ohms, a variation small enough so as not to require an additional isolation network.

The isolation terminating network should also be found useful in reducing ghosts in troublesome installations using unusually long transmission lines. If the antenna transmission line is sufficiently mismatched at the receiver end a portion of the received signal will be reflected back toward the antenna and if the line and the antenna are sufficiently mismatched, as they are sure to be at some frequencies, the reflected signal will be re-reflected back to the receiver. These reflected signals if sufficiently strong will cause multiple ghosts on the picture. Proper termination of the line at either end will eliminate these reflections and the resulting ghosts; a job made to order for Network C since it will maintain proper impedance match at the receiver end of the line for all channels regardless of variations in the receiver input impedance. Of course the signal will have to be sufficiently strong to provide a satisfactory picture after suffering the network loss. Note that



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the losses given in Table 1 are for networks terminated with the rated output (Z2) impedance. This loss will vary with the actual value of antenna input impedance. For instance the loss in Network C will vary from 24.4 to 5.6 db as the antenna input impedance is varied from 30 to 3000 ohms.

Signal Attenuation Applications

Resistor networks may also be used to reduce the signal strength by a required amount while maintaining the proper transmission line impedance termination. Two examples of such applications are discussed below.

In cases of excessive signal strength it is desirable to reduce the signal with an attenuator rather than by rotating the antenna away from the station direction. The attenuator will help to provide a good transmission line termination and will also reduce the susceptability to

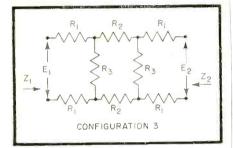


Fig. 3—Attenuation sections in cascade.

ghosts from reflected signals. Station location relative to the receiving location may in some cases dictate the use of a network of suitable attenuation to reduce the signal level from a close signal to within the age range of the receiver while at the same time allowing sufficient signal level from more distant stations.

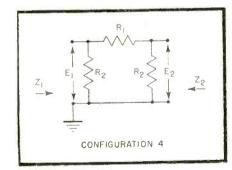


Fig. 4—Attenuation network for use with 72-ohm lines.

Another use for the attenuation network is in service testing. To insure proper operation of sync circuits a test should be made to insure picture sync under weak signal conditions. Resistor networks provide a convenient means for reducing the signal level for such checks. The amount of attenuation may be varied by selection of the proper resistor values to reduce the signal while still maintaining the desired input and output impedances. Network C and D for 300 ohm balanced line provide 10 db and 20 db attenuation respectively. For values greater than 20 db it is desirable to use two attenuation sections in cascade as in Network E. in this arrangement individual resistor valued remain small thus reducing leakage so that the network will provide attenuation at higher frequencies.

Networks F and G provide 10 and 20 db attenuation respectively for use with 72 ohm coax transmission line. Note that with these networks the terminal which is common to the input and output must be connected to the shield side of the transmission line.

Antenna Distribution

When sufficient signal strength is available a resistor network may be used as a convenient means of feeding several receivers from a single antenna. This simple device has the advantages of requiring no power and of operating uniformly over a wide frequency range. Furthermore the inherent insertion loss provides a good degree of isolation be-



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tween receivers so as to minimize the possibility of the different receivers on the same antenna system interfering with each other.

A network arrangement for this application is Configuration 5. Table 2 lists the resistor values and pertinent characteristics for different combinations of impedance and number of receivers in the system. Resistor values have been selected to provide the minimum insertion loss while maintaining proper impedance values. For less than

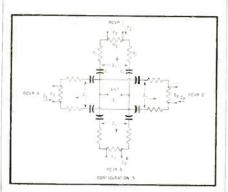


Fig. 5—Network arrangement for feeding several receivers from a single antenna system.

four receiver outlets only the appropriate number of outlet networks of the values shown in Table 2 would be connected. The isolating capacitors, C, are included to prevent power current from flowing between two receivers which have the power and antenna circuits connected to the chassis.

Note that in the Configuration 5 network the isolation between receivers is increased as the network loss is increased that, providing sufficient signal

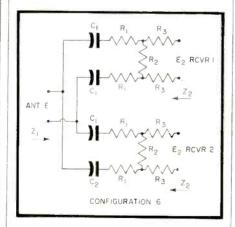


Fig. 6-Two-outlet network with increased attenuation.

strength is available, the increased loss is an advantage in increasing receiver isolation. A two-outlet network with increased receiver isolation is shown as Configuration 6. The values for this arrangement are also shown in Table 2.

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Availability

A number of the networks described herein are available commercially either in printed circuit form or as assemblies of carbon composition resistors in small plastic boxes. Those that are not readily available can be easily assembled in the shop. The accompanying photograph shows one of the 300/300 ohm, 20 db attenuators listed in Table 1 which has been assembled into a small plastic Walsco hardware box.



One of the 300/300 ohm, 20 db attenuators listed in Table 1 assembled in small plastic container.

A four-receiver antenna network, Configuration 5, consisting of parts mounted on a lucite panel in a physical arrangement similar to that shown in the schematic has been found by the author to give good performance. The isolation in this network is sufficient so that disconnecting any number or receivers does not appreciably affect the performance of those remaining connected.

While the networks described herein are basically quite simple devices as evident from the accompanying photograph, care must be used in their assembly to insure best performance. Particularly for use in the uhf TV band, small resistors with very short lead lengths must be used. The very small 1/10-watt carbon composition resistors are well suited to this application since they allow closer spacing and smaller overall size. Extreme care must be taken in the soldering operation to prevent excessive heat from causing damage to or changing the value of the resistors. The heat from the soldering iron applied close to the resistor body can easily cause a permanent increase in resistance of over 10%. Probably the best soldering technique for this application consists of first scraping or sanding the resistor leads and then generously but quickly tinning both the terminals and the resistor leads. Then, after securing the leads to the terminals, and while holding the resistor lead with long nose pliers close to th body, quickly solder each terminal without the addition of any more solder. In this manner the time of heat application is held to a minimum and the pliers help to soak up some heat and keep down the temperature of the resistor body.

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Radio-Television Service Dealer Annual Index for 1954

ANSWERMAN				Television Installation Service Community Antenna System			
44 mc band for TV receivers	June	e p	. 29	Assn., Chicago, Ill	Oct.	p.	22
AC voltage on antennas	. Aug	. p	. 27	Phila. Pa. July p. 41 by Steve Travis	Sept.	p.	21
Admiral Ch. 21J1-width	. Jan.	p	. 32	Texas Electronics Assn Oct. p. 18 Detroit Servicemen's Show Du Mont Chroma Sync Teletron	June	p.	51
Bias Box (troubleshooting agc) Capehart—weak pix	. Nov	. p	. 34	ciation, Kingston, N. Y June p. 53 Entering Hi-Fi, by Gil Tint Enterprising TV Serviceman			
CBS-Columbia Ch. 750-3—height Condenser checker	. July			CIRCUIT COURT Brings Shop to Customer Fifth Annual NATESA Convention	Nov.		
Crosley Model 10-416—audio modu lation of pix		. p	. 36	Capehart CX36—Sync System Mar. p. 52 Get Started in P.A. by Gene Hessel Part 1			
Crosley Model DU-21, Ch. 357-1- focus				DuMont RA160—Vertical Amplifier Mar. p. 52 Sylvania 1-518— Part 2 High Voltage Problems			
Difference between timer and con verter	-			Video Amp., Sync. Keyed AGC., Jan. p. 39 by Paul Goldberg	July	p.	20
Du Mont RA-306-Intermittent pix	ζ.			Horizontal Frequency Troubles by Paul Goldberg	Aug.	p.	36
shrinkage Emerson 600TV—retrace suppres	-			EDITORIAL Horizontal and High Voltage Color Set Circuitry			
sion G.E. 17T103—vertical hold	May May	p p	. 12	A Publisher Hurts Servicemen Oct. p. 5 by Bob Dargan and Sam Marshall Beware of Fake Subscription Part 1		n.	9
G.E. 24C101—ripples in patterns. H.O.T. whistle	. Dec.	p.	. 27	Salesmen Mar. p. 6 Part 2 Changing Service Dealer Trends. Sept. p. 4 Part 3	June	p.	14
Increasing scope gain	Oct.	p.	. 52	Color TV Merchandising Jan. p. 10 Part 4	Aug.	p.	20
Montgomery Ward 2WA—weak Muntz 21" TV—excessive arcing	Nov.	. p.	. 34	Color TV Set Prices Feb. p. 10 Jetomic Antenna Color TV Status July p. 4 by Douglas H. Carpenter by Douglas H. Carpenter		p.	20
Phileo 51-T1606—pix pulls. Phileo Model 1600—no boost de, in	-			Current Production and Sales Figures Sept. p. 4 Key Test Points, by Bob Dargan Part 2	Jan.	p.	26
Sufficient hor, deflection. Philoo Ch. 71RF—vert, deflection.				Instruments and Progress July p. 4 Part 3	Feb.	р. р.	29 33
Phileo Model 48-1001—brightness Phileo field strength meter	Apr.	p.	. 11	Lost Opportunities—Lost Dollars Dec. p. 4 Key Test Points, by Steve Travis New Discounts open to Hi-Fi Part 1			
Proper service techniques	May	p.	18	Servicemen June p. 5 Part 2	Nov.	p.	3
RCA Model 2T51—width Replacing 10BP4 with 10RP4	Aug.	р.	27	New Type Manufacturers Service Policy Nov. p. 5 Lead Dress Problems by Paul Goldberg		p.	32
Replacing tuners with Standard Cascode tuners	Feb.	p.	51	Optimists are We! Jan. p. 10 Oscilloscope Y-Axis Pre-Amplifica- tion, by Harry Casey	June	p.	1.0
Rhombic Antennas Selenium Rectifiers	Mar. July	р. р.	43	Our Prediction is Upheld Oct. p. 5 Performance and Market Analysis Our New Departments May p. 5 of Rotators, by Arne Benson and			
Sylvania Ch. 1-500-1—Horizonta	1			Our New Regular TV Service Information Section Sept. p. 4 Bernard Nussbaum RC Circuits, by Cyrus Glickstein		p.	24
Trav-ler Model 64R50—Intermittent Channel 7 and 13 reception	t			Picture Tube Prices Feb. p. 10 Part 3	May	p.	34
Yagi Antennas on new channels.				Price Cutting is Rampant Feb. p. 10 Part 4 Please Answer Our Questionnaire Part 5	Oct.	p.	30
ASSOCIATION NEWS				Printed Circuit Servicing. Nov. p. 5 Remote Tuners, by San D'Arcy Replacing Printed Circuit			
ASSOCIATION NEWS Associated Radio & TV Servicement				Random Notes Feb. p. 10 Components, by Admiral Corp Sectionalized Printed Circuit	Nov.	p.	7
of Chicago	June	p.	53	Scarce Metal Shortage	Dec.	р.	19
Associated Radio-TV Service Deal- ers of Columbus, Ohio		p.	58	Shop Efficiency May p. 5 Sell and Service Radio-Controlled Starting Our Fifteenth Year Apr. p. 5 Garage Loor Openers			
	Apr. Sept.			The Customers Write Jan. p. 10 by San D'Arcy	July	p.	33
Associated Radio-TV Servicemen				The Readers Write June p. 5 Servicing Marine Radio, Part 1 TV Channel Changes Feb. p. 10 by Marvin Klein	June	p.	24
Certified Electronic Technicians Assn., New York				TV Receiver and Station Count. Dec. p. 4 Servicing Video I-F Systems We Apologize Mar. p. 6 (Symposium Series No. 10)			
Council of Radio & TV Service				We Caused a Furore Sept. p. 4 by Matthew Mandl Servicing Vertical Instability	Feb.	p.	21
Associations, Penna	Mar.	p.	58	Worthwhile Income Booster Ideas May p. 5 You Ask For It May p. 5 Smello-Vision, by Oscar Fisch Smello-Vision.	Oct.	p.	10 52
	Apr. July			Solving Sweep-Alignment Problems by Robert G. Middleton			
Electronic Service Association De- troit, Mich.	July	D.	41	FEATURE ARTICLES Part 1 Part 2	Mar.	p.	27
Empire State Federation of Elec- tronic Technicians Assns., N. Y.				A.G.C. Fronteins, by Fatt Goldberg. July p. 20 Part 3	Aug.	p.	10
Eastern TV Conference	Mar.	p.	17	A New Cathode Ray Tube Tester Servicing Remote Tuners			
	June	p.	41	by Morton Greenberg Nov. p. 49 A New License Bill Stand-off Insulators and UHF			
Federation of Radio Servicemen's	Sept.			by Abraham Frankel July p. 24 A New Deal For Hi-Fi by Carpenter and Tolins Sync Amplifier Problems	Jan.	p.	40
Associations, Penna	Jan. Sept.			by Noel Namtro	Oct.	p.	34
Long Branch Radio Technicians Association. Long Branch, Cal.	Apr.	D.	30		Dec.	p.	34
Long Island Electronic Technicians Association, L.I., New York				A New All-Channel VHF-UHF by Robin Blanchard	× 1.		1.4
	Feb. Mar.	p.	56	by Harris and Greenberg Apr. p. 53 Torture Test for TV Rotators			
Long Island TV & Radio Techni-	Apr.			Applications of Resistive Networks by J. C. Geist Dec. p. 31 by Herbert G. Koenig Transistor Portable, by Regency	Mar. Dec.	p. :	24
				A Revolutionary Method of True and False Color Concepts Antenna Coupling by Matthew Mandl	July	p.	18
	Apr. June			by Harold Harris	Nov.	p.	50
	Sept. Oct.			by Jacob Anthes. July p. 22 CBS Columbia Ch 750-3, 751-3	Oct. Nov.		
National Alliance of Television and Electronic Service Associations				Audio Adaptations for Profitable CBS Columbia Color TV Receiver			
	Mar. Sept.	p.	58	Auto Inverters, by San D'Arcy June p. 23 TV Colorimetry			
National Electronic Technicians &				Block Diagram Analysis of Color Transmission and Reception TV Instrument Clinic	Apr.	р. :	21
Service Dealers Assns	Oct.			by Bob Dargan and Sam Marshall Part 1 Sept. p. 9 Part 1 Part 1	Apr.	р.	12
Philadelphia Radio Servicemen's Association, Inc.	Feb.	p.	16	Part 2 Oet. p. 14 Part 2 Nov. p. 11 Part 3 Part 3	May	p. 4	41
	Apr. Nov.	p.	8	CBS Color Tube	Sept. 1	1). 2	29
Radio & TV Association of Spring- field, Ohio	Mar.			Receivers Part 6	Dec.	p. 2	23
Radio & TV Technicians' Guild, Gadsden, Alabama	June			Color Oscillator and Central Circuits zontal Output Transformer Cir-			
San Antonio Radio & TV Associa-				Color Stripe Signal May p. 29 Part 1			
Southern Pennsylvania Radio & TV	Mar.			Part 3 Jan. p. 23 Universal Test Bench Connectors	Dec. 1	p. 1	16
Technicians Association	Apr.	p.	45	Part 4 Feb. p. 25 by Irving Tepper	Mar.	р. З	31

Video Amplifier Servicing Hints by Robert G. Middleton	Mar.	p.	27
Waveform Analysis in Sync Circuits, by George Holmes	Jan	p	45

GUEST EDITORIAL

Edward F. Mo					
H. J. Schulma	ın, Du	Mont	Feb.	p.	12

NEW PRODUCTS

Jan. p 46, Feb. p. 54, Mar. p. 56, Apr. p. 54, May p. 51, June p. 42, July p. 34, Aug. p. 34, Sept. p. 52, Oct. p. 50. Nov. p. 52-54, Dec. p. 48.

RIDER TV FIELD MANUAL SERVICE DATA SHEETS

SEKVICE DATA SHEETS			
Admiral TV Chassis 22C2: 22E2	Mar	13.	43
	Sept.		
Arvin 1900 Series	May		
Arvin 1900 Series	3.000,	10.	
CT-74	Mar.	р.	47
Dumont RA-166, -167, -170, -171	Nov.		
732B, 734B, 733F	Feb.	p.	43
Fada Models 17C2, 17C4, 17C6,			
17T9 Firestone Models 13-G-128, -129,	May	p.	47
-130, etc.			
Codes 334-3-MS39A, 334-3-MS39B.	Oct.		
Firestone Model 13-G-135	Nov.	p.	43
General Electric Models 17C125.	-		4 =
20C107, 20C201, etc.	Sept.	p.	47
Hallicrafters Models 1067,1068,			0.0
Chassis L1100D, runs 1 & 2	June	p.	55
Hoffman, Chassis 400. Models 21M721, -B723M722 and			
Chassis 401, Models 21M146,			
-B147, -P148, -M320, -B321,			
	Taria	n	95
-P322 Magnavox Chassis CT,	a trix	ρ.	4.)
Series 331 through 357	Anr	p.	
Majestic, Series 112-17", 113-21"	Tuly	p.	
Motorola, Models 17F13C etc	June	p.	37
Olympic, TM-TN Chassis	Nov.	D.	45
Pacific Mercury Chassis	2.077	1	
200-1, 2, 3, 4, etc.	Oct.	p.	39
Philco Ch. R-191, D-191, Code 140	Dec.	p.	43
RCA KCS74 and KCS74M1	Jan.	p.	41
RCA KCS81, Models 305, 305U,			
317, 317U, etc	Aug.	p.	31
Sentinel, Mods. LU500, 510, 511,			~ ^
etc.	Aug.	p.	29
Stewart-Warner, Model 9132A	Dec.		
Sylvania Chassis 1-508-1, 1-508-2	Feb.	p.	47
Western Auto Model 2D-1315A,			
2D1325A, etc., Ch. 21T1A, -2A,	0 .		4.1
17T1B Westinghouse, Ch. V-2233-1, Models	Oct.	p.	4.1
H740T21, H742K21, H743K21	Sout	73	10
ALITORES, ALITERES, III TORES.	sept.	Q.	40

TRADE FLASHES

Jan. p. 6, Feb. p. 6, Mar. p. 8, Apr. p. 6, May p. 52, June p. 50, July p. 42, Aug. p. 42. Sept. p. 16, Oct. p. 54, Nov. p. 20, Dec. p. 50.

TRADE LITERATURE

Feb. p. 16, May p. 63, June p. 56, July p. 45, Oct. p. 58.

VIDEO SPEED SERVICING SYSTEMS

AIDEO SLEED SEKAICING 2	1216	M:	•
Admiral 21B1	Nov.	p.	35
Crosley Model 10-419MU	Sept.	p.	37
Capehart CX-33DX	June	p.	45
Du Mont chassis RA109	Feb.	p.	35
Du Mont chassis RA105	Apr.	p.	37
Du Mont chassis RA164	June	p.	47
Du Mont chassis RA166	July	p.	29
Du Mont chassis RA306	Aug.	p.	23
Emerson chassis 120066-B,			
Mod. 571. General Electric 830 "R" Version.	Mar.	p.	
General Electric 830 "R" Version	Jan.	p.	35
Hallicrafters, Mod. 805, ch.			
M800S	Apr.	p.	39
Hallicrafters, Mod. HA-861-1	Sept.	p.	39
Kaye Halbert Model 231	July	p.	31
Majestic Model 121B	Dec.	р.	37
Motorola, ch. TS32B, Mod. 12VT13	Mar.	p.	37
Muntz 2053A, chassis 17B1 Phileo Mod. 50-T1600, 51-T1600.	Nov.	p.	37
	D-0		39
Phileo chassis R181, D181	Dec. Oct.	р. р.	43
Phileo chassis R191, D191	Aug.	D.	25
Phileo RF41, 42, 43	Feb.	р.	37
RCA KCS47A	May	p.	37
RCA KCS68C	Nov.	D.	39
RCA KCS81	Dec.	p.	41
Sentinel Model 1U425	Sept.	p.	41
Sylvania 1-504-1	May	p.	39
TeleKing 812	Jan.	p.	37
Westinghouse chassis V2150-101	Mar.	p.	39
Zenith 23G22	Feb.	p.	39
Zenith Chassis 21L21	Oct.	p.	48

WORKBENCH

WOKKBEITCH			
Admiral 19B1-			
Shift in Oscillator Frequency	Nov.	p.	
Admiral 20V1—Ruzz	May	p.	30
Admiral 20V1—Buzz Admiral 21B1—No Brightness Admiral 21C1—Lack of Horizontal	July	p.	20
Admiral 2101 Lask of Horizontal		•	
Admiral 2101—Dack of Horizontal	Dec.	p.	34
and Vertical Hold	June		41
Admiral 22M1—Horizontal Drift	Aug.		36
Admiral 22M1—Horizontal Drift Admiral 22M1—Smeared Pix	Sept.		33
Admiral 22M1—Smeared Pix	sept.	1,.	00
Capehart CX-33-	Feb.	~	33
Horizontal Wiggle	reb.	p).	00
Capehart CX-33-	- 1		20
Intermittent Blooming	July	p.	20
Crosley Mod. 10-423MX-			
Short in Front End	Nov.	p.	51
Du Mont RA-103C-Horizontal			
Frequency Out of Range	Aug.	p.	36
Du Mont RA-103C-Vertical Roll			
and Horizontal Wobble	Dec.	p.	35
Du Mont RA-103D-			
Insufficient Height	Jan.	p.	33
Du Mont RA105—	0		
Pix and Sound Fades Out	Sept.	n	32
Pix and Sound rades Out	Debr.	1.	
Du Mont RA164-	Mar.	~	34
Critical Vertical Hold	W1 24 I -	ρ.	0.
Emerson Mod. 654B, ch. 12011B-			5.1
Video Overload	.June	р.	23 1
Emerson Ch. 120162A-Horizontal			0.4
Pull and Vertical Roll	Oct.	p.	0.0
Freed Model 55-Hum	Sept.	p.	.).)
General Electric 10T1-			
Critical Vertical Hold.	Mar.	p.	34
Horizontal Tearing Motorola TS-60—Loss of Channels Motorola TS-60—	Feb.	p.	
Motorola TS-60-Loss of Channels	Jan.	p.	34
Motorola TS-60-			
Poor Horizontal Hold.	Feb.	p.	33
Motorola TS95-Sound Distorted at			
High Volume Setting	May	p.	31
Motorola TS114-	171. 61.7	· Lo	
MOIOFOIR ISI14-			
Intermittent Horizontal Pull	Oct.	n	25
and Vertical Jitter	Oct.	,,.	170
Philco Model 52-1882-121,			
RF chassis 44. Defl. ch. D4—	37		90
Vertical Roll Philco 1954 "B" TV181—	Mar.	p.	00
Phileo 1954 "B" TV181-	-		0.1
1B3's Burn Out	July	p.	21
RCA KCS 47A—			
Intermittent H.V	Jan.	p.	
RCA KCS82—Horizontal Pull.	Oct.	p.	34
Intermittent H.V. RCA KCS82—Horizontal Pull. RCA 630TS—Pulling at Top of Pix			
and Vertical Slipping	Dec.	p.	35
Tele-King 812-Distortion	May	p.	30
Westinghouse chassis 21050-01-		-	
No Raster	July	p.	21
Zenith, Ch. 24G26-			
Low rf Gain	Nov.	p.	50
The of Charles and the second		P .	.,

ASSOCIATION NEWS

[from page 8]

the problem of handling sets through an intermediary who later claims the set was not his so he wants the set returned without charge. Storage of these "orphans" and protection against losses pose serious and costly problems.

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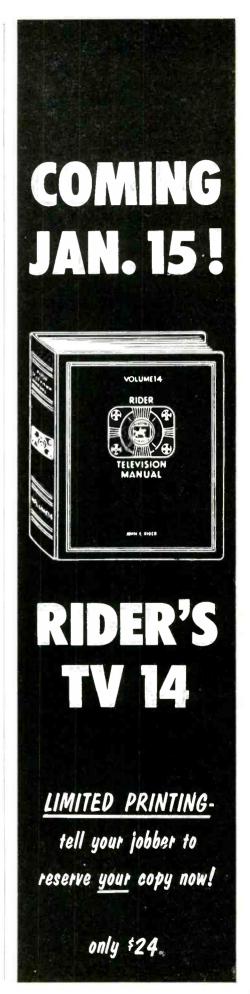
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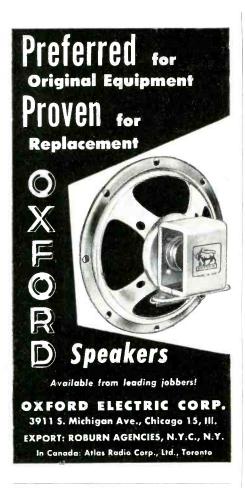
In consenting to the removal of a coceptance for service of the unit described on this receipt, the undersigned states that it is his har property and that service is being assumed by the company with the understanding that the set will be stored for a period of not more than 30 days following notice of completion of service, phone or post card notice is deemed sufficient. After 30 days, it agreed that a storage and protection fee of 25c per day will be added to the billing. It is further agreed that should the goods not be claimed after 60 days from date of this receipt and should no other previous arrangement in writing be agreed upon, the company has the authority to junk, sell or in any other way dispose of the goods and be completely released of any and all claims against it orising from the transaction covered by this receipt.

Set Owner Signed Company Date

RTSA—Pittsburgh

The RTSA of Pittsburgh, Inc., is in the process of setting up a course in TV Training to upgrade the qualified technician. This subject was brought up the last regular meeting, and quite some enthusiasm was shown by the show of hands.





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Advertising Index

Acme Electric Corp.	57
Astron Corporation	18
Audel Publishers	64
Authorized Manufacturing Co	61
Belden Manufacturing Co.	2
Blonder-Tongue Laboratories, Inc	52
Bussmann Manufacturing Co.	
Channel Master Corp. 6	, 7
Chicago Standard Transformer Corp.	49
Clear Beam Antenna Corp.	53
Columbia Wire & Supply Co	33
Cornell-Dubilier Electric Corp9,	58
Electric Soldering Iron Co., Inc.	61
r = prine	, ,
Jensen Industries	
Jensen Manufacturing Co.	
JFD Manufacturing Co., Inc.	60
Lowell Manufacturing Co.	55
Miller, J. W. Co.	54
Ohmite Manufacturing Co	8
Oxford Electric Corp.	
Oxford Electric Corp.	04
Precision Apparatus Co., IncCover	3
Radiart Corporation	9
Radio City Products Co., Inc.	
Raytheon Manufacturing CoCover	2
RCA Tube DeptCover	4
RCA Service Co., Inc.	56
Remot-O-Matic Sales, Inc.	14
Rider, John F. Publisher56,	63
Sams, Howard W. & Co.	
Seco Manufacturing Co.	
	22
	64
Standard Coil Products Co., Inc	21
Transvision	64
Tung-Sol Electric, Inc50,	- 4
United Catalog Publishers	53
Walco	60
	26
Weston Elec. Instrument Corp.	5



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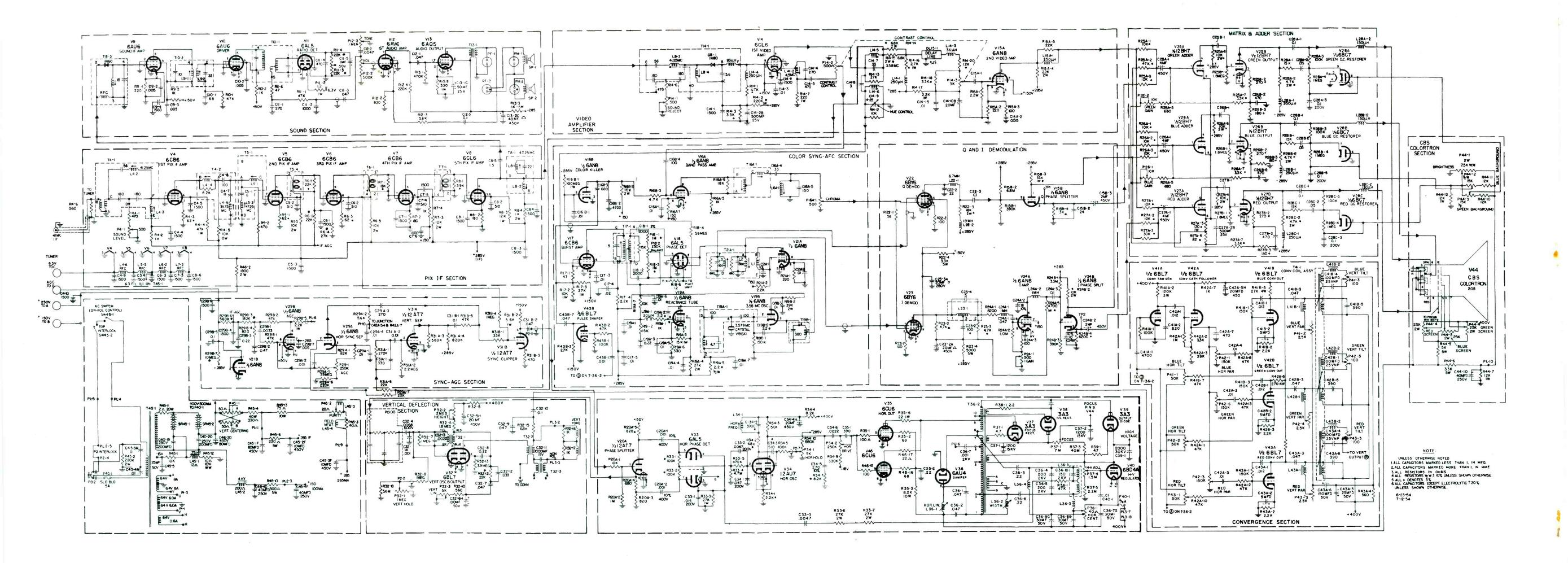
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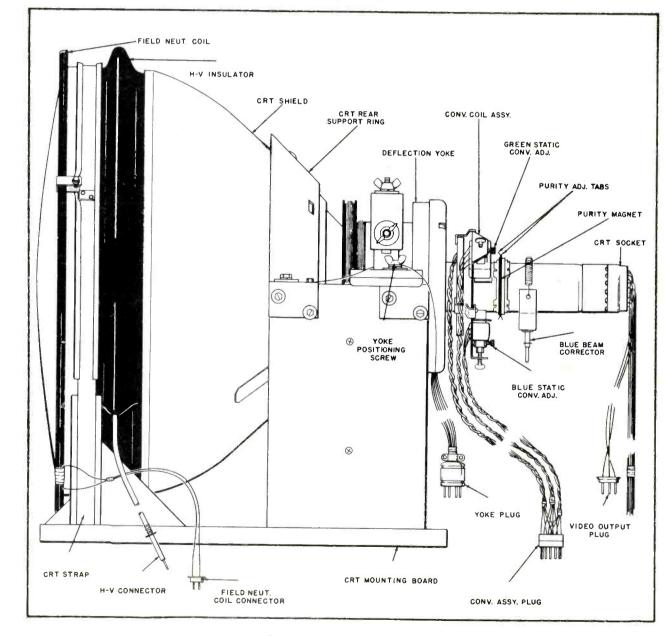
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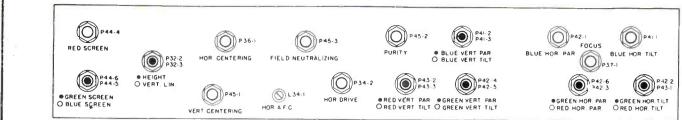
Picture tube assembly.

Specifications

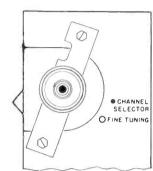
1						Specifications
Picture Tube				Tube Complement		
Туре		CBS Hytron 19VP22				
Dimensions		19-inch round, all glass. Narea 205 square inches		Symbol	Type	Function
	•	nea 200 square n	iches	V1	6BZ7	RF Amplifier
Operating Contro	ole e	S Station Selector, Fine and Ul		V2	6N8	Mixer and Oscillator
Operating Contro		Funing, Vertical		V3	6AF4	UHF Oscillator
]	Hold, Contrast,	Volume, Tone,	V4	6CB6	1st IF
	•	Chroma and Hue		V5	6BC6	2nd IF
D				V6	6BC6	3rd IF
Power Rating		05 100 1 60	1 40	V7	6CB6	4th IF
Consumption	5	105—120 volts 60-cycle AC 500 Watts		V8	6CL6	5th IF
•	·			V9	6AU6	Sound IF Amplifier
Number of Tubes		41 tubes plus 3 rectifiers, 2 seleni- um rectifiers and 3 crystal diodes		V10	6AU6	Driver
				V11	6AL5	Ratio Detector
				V12	6AU6	Audio Amplifier
Ant. Input Imped	dance			V13	6AQ5	Audio Output
VHF		300 ohms balanced		V14	6CL5	1st Video Amplifier
UHF	3	300 ohms balanced		V15	6AN8	2nd Video Amplifier-Q Phase Splitter
F				V16	6AN8	Band Pass Amp.—Color Killer
Frequency Range		Channels 2 through 82		V17	6CB6	Burst Amplifier
			V18	6AL5	Phase Det.	
Intermediate Fre				V19	6AN8	Reactance Tube-3.58 mc Oscillator
Video IF Sound IF		5.75 mc 1.25 mc		V20	12AT7	Horizontal Phase Splitter
Sound Intercarrier Fr	eq.	4.5 mc		V21	6AN8	1 Amplifier—AGC Clamp
Color Subcarrier Free		2.17 mc		V22	6BY6	Q Demodulator
Adjacent Channel Son Trap Freq.		9.75 mc		V23	6BY6	I Demodulator
Adjacent Channel Vid	leo			V24	6AN8	I Amplifier—I Phase Splitter
Trap Freq. Accompanying Channel Sour for less than 10% distortion		7.25 mc		V25	12BH7	Green Adder-Green Output
		41.25 mc		V26	12BH7	Blue Adder-Blue Output
,,,				V27	12BH7	Red Adder-Red Output
Crt High Voltage		26 kv adjusted		V28	6BC7	Green Red Blue DC Restorers
				V29	6AN8	AGC Amp-Horiz, Sync Separator
Loudspeakers				V31	12AT7	Vert. Sync Separator—Sync Clipper
Size	т	wo, 6½ in x 9¼ i	n sneakers	V32	6BL7	Vert. Oscillator-Vert. Output
Voice Coil		3.2 ohms at 400 cycles		V33	6AL5	Horizontal Phase Det.
				V34	12AU7	Horizontal Oscillator
Focus		Electrostatic		V35.	6CU6	Horizontal Output
				V36	6AU4	Damper
Deflection		Electromagnetic		V37	3A3	HV Rectifier
				V38	3A3	HV Rectifier
Convergence				V39	3A3	HV Rectifier
Static		PM Magnetic Electromagnetic		V40	6BD4	HV Regulator
Dynamic	E			V41	6BL7	Convergence sawtooth Gen-Blue conv. Output
Cabinet Dimensions				V42	6BL7	Green conv. Output—Conv. Cathode Follower
Model	Width	Height	Depth	V43	6BL7	Pulse Shaper-Red. Conv. Output
205C1 Console	341/4	42-11/16	26-1/16	V44	19VP22	Colorton CRT
205C2 Console w/Doors	341/4	42-11/16	26-15/16	V46	6CU6	Horizontal Output

Warning - High Voltage

POTENTIALS AS HIGH AS 26,000 VOLTS ARE PRESENT WHEN THIS RECEIVER IS OPERATION. OF THE RECEIVER OUTSIDE THE CABINET OR WITH COVERS REMOVED INVOLVES A SHOCK HAZARD FROM THE RECEIVER POWER SUPPLIES. WORK ON THE RECEIVER SHOULD NOT BE ATTEMPTED BY ANYONE WHO IS NOT



Deflection chassis front controls.



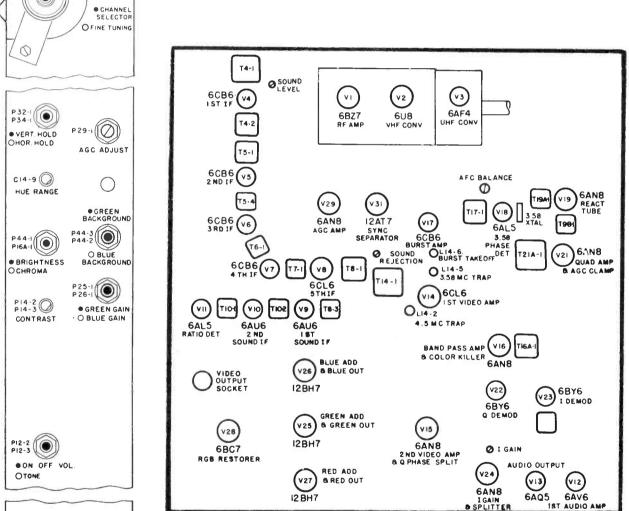
C14-9 HUE RANGE

BRIGHTNESS OCHROMA

P14-2 P14-3

ON OFF VOL.

OTONE



Signal chassis front controls.

Top view of signal chassis

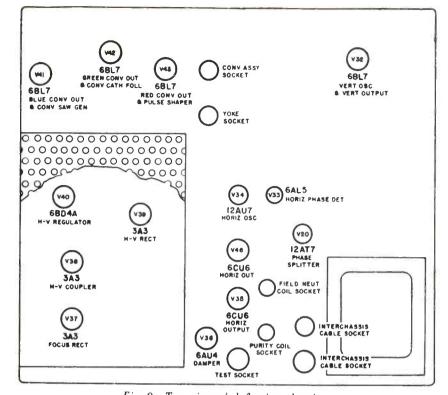
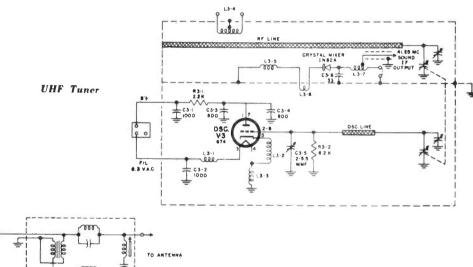
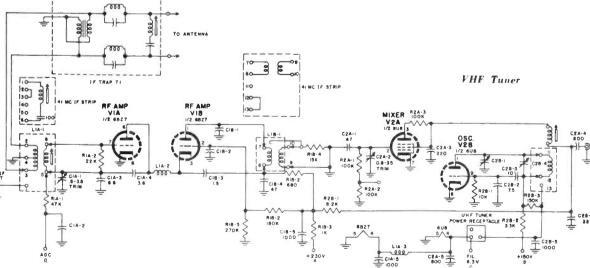


Fig. 9—Top view of deflection chassis





PRECISION



GENERAL SPECIFICATIONS:

* VARIABLE-FREQUENCY SINE-WAVE RANGES:

for testing audio amplifiers, low frequency RF amplifiers, etc.: Continuous Coverage from 20 Cycles to 200 Kilocycles in Four Bands.

* VARIABLE FREQUENCY SQUARE-WAVE RANGES:

for analyzing audio amplifiers, wide-range amplifiers, etc.: 20 Cycles through 20,000 Cycles in Three Bands.

* FOUR FIXED, HIGH-FREQUENCY SQUARE WAVES:

for analysis of video and other wide-band amplifiers up to 20MC band-width: 50~KC-100~KC-250~KC-500~KC steps.

* OUTPUT CHARACTERISTICS:

Variable Frequency Ranges: 0.2000 ohms, 0.10 volts RMS, flat within ± 1 db. Accuracy: $\pm 2\%$ from 50 cps. to 200 KC. ± 1 cps. from 20 cps. to 50 cps. Distortion: Less than 1% from 20 cycles through 200 KC. 20 KC Square-Wave Rise Time: .5 microseconds.

Fixed High Frequency Square-Waves: 0-250 ohms, 0-5 volts P-P Rise Time: .05 microsecond • Overshoot: Negligible

- **★ TUBE COMPLEMENT:** 1-5879, 1-6CL6, 1-6J6, 2-6AU6, 1-6BL7, 1-6AH6, 1-6X4.
- ★ SEPARATE OUTPUT CIRCUITS: for the variable and fixed frequency ranges. Dual pilot lamps automatically indicate the active output jacks.
- ★ TERMINATED, LOW-LOSS, HIGH FREQUENCY COAXIAL OUTPUT CABLE:

transmits the H.F. square waves to circuits under test, without distortion.

* EXTERNAL 'SYNC' TERMINAL POST:

for synchronizing oscillograph horizontal sweep to H.F. square-wave.

★ ETCHED-ANODIZED TUNING DIAL and PANEL: NO-glare, engine-turned dial finish and soft-black panel field afford utmost visibility and ease of reading.

MODEL E-300: in black, ripple finished, portable steel case $-10\frac{1}{2}x12x6$ ". Complete with tubes, coaxial output cable and operating manual.

Net Price \$17500

Announces a NEW BASIC TEST INSTRUMENT

- ...for Laboratory and Test Bench
- ...for Engineer and Technician

THE MODEL

E-300

SINE-SQUARE WAVE SIGNAL GENERATOR

(AUDIO-VIDEO RANGE)

THE NEW SERIES E-300 has been especially developed to answer many modern electronic amplifier testing problems which cannot be handled with just the usual complement of test instruments.

The Series E-300 provides accurate sine and square wave signals for direct performance testing of:

High Fidelity Audio Amplifiers TV Video Amplifiers Carrier Current Systems

... and other wide range devices, etc.

Sine-Square Wave Analysis, with the Series E-300, streamlines amplifier test procedure and assures more uniformly high standards of apparatus performance, because sine-square wave testing is a most reliable indicator of:

Frequency Response
Phase Shift
Amplitude Distortion, etc.

The operating Manual for the Series E-300 has been especially prepared to describe the basic techniques of sine-square wave testing. The information establishes a foundation that will permit the operator to interpret sine-square wave-forms in terms of frequency response, distortion, etc.

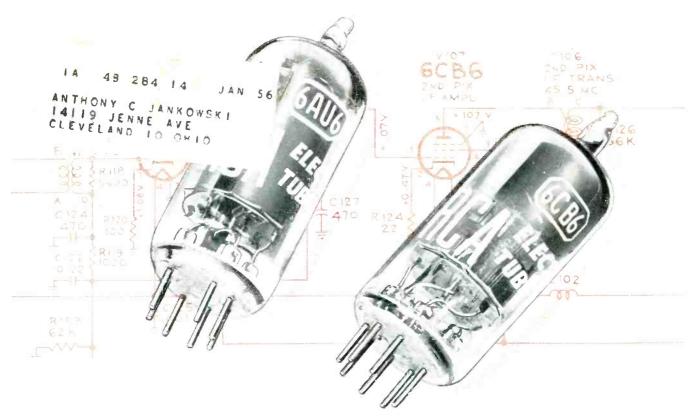
You may obtain this comprehensive Manual, at only 25¢ per copy, to cover cost of printing and handling. Write directly to factory.



PRECISION Apparatus Company, Inc.

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