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EDITORIA

ITH over 100 models of some W forty manufacturers now on the production lines, after having been price approved by the OPA, we find that the postwar program is at long last under way. While receivers are not being made in any great quantities because of parts shortages and labor problems, it is promising to note that there is production, production that we're certain will be stepped up during the next few months.

Most of the receivers made thus far have been of the a-m, five to six-tube-table model type, with $4\frac{1}{2}$ " to 6" speakers. Some of the receivers are featuring automatic phono operation with crystal pickups. Several push-button models have also been made, featuring six and eight-button controls. The a-c/d-c/battery models have included self-charging switches, a feature that will probably be included in many house line-battery types next year. In the main, the models are covering the broadcast band. A few have included the short-wave band. About a dozen console models also have been announced. These are all a-m, with from seven to eight tubes, and with phono systems.

systems. To facilitate installation and maintenance, several manufacturers have announced that they will include simplified servicing notes with every receiver. Where the models become complex, simplified tracer circuits are planned. Incident ally, many of the multi-tube multi-system re-ceivers will actually be less complex than the prewar models, due to the use of new combina-tion tube types, simplified circuit systems and smaller components. However, the standards will be necessarily higher because of the higher gain and improved fidelity provided. Thus more care than ever will be exercised in manufacture, a care that will demand close scrutiny by every Service Man. It will be up to the Service Man to maintain these standards in repair and main-tenance.

A thorough knowledge of circuits and com-plete familiarity with component design and fabrication are musts in the postwar receiver servicing program. Continuous study and read-ing of such articles that appear in SERVICE are essentials in this program. Returning G.I. Ser-vice Men can brush up on radio servicing by attending special courses now being given at trade schools throughout the country. In New York, for instance, the Manhattan Trade Center offers complete basic and refresher courses that will prove quite handy in starting up the right wav. A thorough knowledge of circuits and comwav.

way. Service Men will, in 1946, also find many new fields to service. We've described some of these in past editorials . . . aircraft, apartment houses, plants, etc. A particularly lucrative field is marine craft. With thousands of large and small boats soon to appear with receivers and trans-mitters, there'll be an urgent call for Service Men. As in aircraft, commercial licenses will be required to operate and then service the equipment. But the extra effort required to secure the license will pay substantial dividends. And we must not forget the radio installations in schools, shops, autoshop restaurants and dozen of other businesses using radio, that offer excellent installation and servicing opportunities for the Service Man. 1946 should be a profitable and active year

1946 should be a profitable and active year for every active Service Man!

A Monthly Digest of Redio and Allied Maintenance

Reg. U. S. Patent Office

Vol. 14, No. 12

ALFRED A. GHIRARDI

Advisory Editor

December, 1945

LEWIS WINNER

Editorial Director

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James C. Munn, 10515 Wilbur Avenue, Cleveland 6. Ohio: Telephone SWeetbriar 0052 Pacific Coast Representative: Brand & Brand, 1052 W. Sixth St., Los Angeles 14, Calif.; Telephone Michigan 1732

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SYLVANIA NEWS RADIO SERVICE EDITION

DEC. Published by SYLVANIA ELECTRIC PRODUCTS INC., Emporium, Pa. 1945

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AUTOMATIC NOISE L I M I T E R S

ANY circuit improvements featured in the better type home receivers are also included in the communications type receiver. However, because of the special applications of the communications receiver, these circuit designs usually have to be expanded. The squelch or automatic noise limiter (a-n-1) is an excellent example of such treatment. In home s-w receivers, noise limiters are used to reduce static and local noise interference. Since some limitation is usually imposed on tone quality by a-n-l use, this feature is usually not used in home receivers designed for b-c reception alone. In communications receivers, which are used for communications in the ham bands the automatic noise limiter sometimes makes the difference between a readable and unreadable signal.

Most a-n-l systems operate to silence the receiver momentarily upon

by THOMAS T. DONALD

the receipt of a sharp, high-amplitude signal, such as we encounter in static and other sharp noises. Transient noises, such as ignition interference, have steep wave fronts, occur momentarily with relatively long pauses, and consist mostly of high audio-frequency components. Thus systems have been designed which sharply attenuate these high-amplitude, high audio-frequency, short-duration components, and thereby facilitate signal reading.

Basic Circuits

Three basic circuits for momentarily quieting a receiver are shown in Fig. 1 a, b, c. The a-n-l circuit of Fig. 1a consists of a noise amplifier and noise rectifier. In operation, any signal fed to the grid of the controlled i-f tube, a 6L7, is also fed to the grid of the noise amplifier. The bucking bias voltage on the cathode of the noise rectifier is set to the amplitude of the desired signal. Any sudden noise input voltage of appreciably greater amplitude than the desired signal, will be rectified by the 6H6 noise rectifier. This rectified

Fig. 1 (above). Three basic a-n-l circuits. In a, a noise amplifier is used in conjunction with a noise rectifier to momentarily blank out the signal by overbiasing the i-f amplifier. A series diode type is shown in b. By setting the bias of the second diode to the level of the desired signal, any voltage, such as a sharp noise pulse, in excess of the desired signal, will be clipped. Fig. 1c shows an automatic type of noise control which blanks out any noise signal in excess of 100% modulation.



noise voltage, by biasing the second control grid of the 6L7, will momentarily bias the tube to cutoff, thereby rendering the receiver inoperative for the duration of the noise signal. The effectiveness of the circuit is controlled manually by adjusting the bucking bias voltage applied to the cathode of the noise rectifier. The circuit operates on the principle that most noises are of short duration, with relatively long intervals between pulses. Since the quiet period of the receiver is only a few microseconds, there is no apparent discontinuity in the received signal.

Series-Valve Noise Limiters

Fig. 1b shows a second type of noise limiter operating at the second detector stage. This is known as a series-valve noise limiter. The operation of this circuit depends on the conductivity of the second, or limiting diode. Its conductivity depends on the plate maintaining a positive relationship to the cathode. Any signal detected by the first diode will create a voltage, across R₁, of negative polarity with relation to ground. Since this negative voltage will vary depending on the modulation of the received signal, and since it is applied to the plate of the second diode through the coupling capacitor, the net voltage on the plate of the second diode will be a function of this voltage and the posiFig. 2. The a-n-l circuit of the Hallicrafters SX 28. This circuit is identical in operation to that shown in Fig 1a. The ave amplifier is used to amplify the noise voltage as well as the ave voltage.

tive plate voltage applied from the power supply. (This action may be considered as a form of voltage modulation.) As long as the negative voltage does not exceed the positive plate voltage, the second diode will conduct. Therefore, by setting the arm of R₂ to a voltage level which is equal to the amplitude of the desired signal, the sudden, sharp, high-amplitude noise impulses are clipped.

Automatic Control

Fig. 1c shows a third type of noise limiter which needs no manual control, since it automatically adjusts itself to the signal. The value of R_t is so chosen that the voltage on the control grid of the 6N7 is double the cathode voltage of the same tube, with relation to ground. Since the plate of the 6N7 returns to ground, this is equivalent to putting a positive voltage on the plate of the tube, and a negative voltage on the control grid. Since the voltage on the plate is of the order of +4 volts, and that on the

grid -2 volts, the tube may be considered as being non-conductive. R_4 and C_1 constitute a filter network which removes the audio-modulation component of the received signal from the 6N7 control grid. However, the cathode voltage, or by inverse relationship, the positive plate voltage, will follow the amplitude of the audio modulation. For values up to 100% modulation, the cathode voltage will approach, but never drop below the control grid voltage. However, for modulations in excess of 100%, the cathode voltage will drop below the grid voltage, which is biased by the carrier alone. Under these conditions, the grid becomes positive with relation to the cathode, and the tube becomes conductive. The 6N7 is chosen, and operated in parallel, since it has a very low plate-to-cathode impedance. Since the plates of the 6N7 return to ground a low impedance shunt load is imposed across the audio input.

Hallicrafters SX28

Most communications receivers use some form of one of these three basic a-n-l circuits. Figure 2 shows an adaptation of Fig. 1*a* as used in the Hallicrafters SX28. A 6B8 serves a dual purpose, in that it acts as both a noise amplifier and a avc amplifier. The a-n-l control, instead of controlling the bucking bias on the noise

rectifier alone, also controls the gain of the second noise amplifier, 6K7, by increasing the bias on this tube and, at the same time, the bias on the 6H6 noise rectifier. Another feature is the high audio-frequency characteristic of the 6H6 noise rectifier. By using a balanced-type transformer to feed the noise rectifier, very little of the i-f frequency appears across the detector load circuit. This permits th use of a low-pass filter. This lowpass filter permits the passage of high audio frequencies without attenuation, for biasing the second input grid of the 6L7. Since, as stated previously, high audio frequencies make up the greater portion of noise signals, the effectiveness of the limiter is thereby increased for the higher frequencies, through the application of higher voltage levels to the second 6L7 grid for high audio-frequency noise components.

Hammarlund SP-200X

The a-n-l system used in the Hammarlund SP-200X (Fig. 3) is similar to the one shown in Fig. 1c. A resistor and capacitor has been added to increase the effectiveness of the circuit. As previously stated, noise pulses are of very short duration. For this reason a very fast time constant is used in the RC network in the cathode of the 6H6 detector, about .0000125 second. This same network also improves the operation of the a-n-l, since at the same time that the grid goes positive with relation to the cathode, a higher plate voltage is applied to the plate of the 6N7, thereby increasing its conductivity.

National NHU

Fig. 4 shows the a-n-l system of the National NHU communications receiver. This circuit is similar to that shown in Fig. 1b. However. here an infinite impedance detector is used, necessitating some changes in the circuit. Since the two sections of the twin triode are cathode coupled, the signal voltage applied to the cathode of the noise limiter stage is positive. Ground, in this case, is positive with relation to B—. The grid and plate are hooked up in a semi-diode circuit, in that the grid is positive with relation to cathode, as well as the plate. A sharp noise pulse of voltage in excess of the plate-grid voltage will place the cathode at a higher potential than the plate-grid. This renders the tube non-conductive for the duration of the noise pulse. The time constant of the

(Continued on page 46)

adanta Fig. 3. An adapta-tion of the circuit of Fig. 1c; Hammarlund SP 200%. The RCnetwork in the plate circuit is used to im-prove the effectiveness of the c a l action



Fig. 1 An

of the a-n-l action.

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Fig. 5. The a-n-l cir-cuit of the RME 41-43. This circuit is similar to that shown in Fig. 1c. The in-sert shows how this circuit may be inter-preted as a Wheatstone Reide type of circuit Bridge type of circuit.



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Fig. 1. Three methods of reducing interference from strong local signal transmissions. In a is shown a parallel type of wave trap familiar to most Service Men. Parallel is used in the sense that the trap is in parallel with the antenna coil. In b a series type of wave trap is shown. Note that intrinsically, the wave traps are the inverse of their position in the circuit; a series type trap is used in parallel, and vice versa. Fig. Ic shows how the primary of an antenna coil may be adapted to a wave trap by adding a padder capacitor in series with the primary. Since most receivers use high-impedance primaries, a small padder must be used to tune in the b-c range. Otherwise turns must be removed from the primary to bring the circuit within resonant range.

CROSS MODULATION BEAT NOTES and R-F WHISTLES

THE signal - interference and whistle complaint has always been a provoking problem for the Service Man. Often the trouble is further complicated by the fact that the interference may be local, or caused by the receiver itself. Thus it becomes necessary to evaluate the source as well as the remedy.

Studying the Problem

To study the problem, an experiment can be conducted on superheterodyne receivers, preferably those with two-ganged tuning capacitors. First a r-f signal generator should be con-

by A. M. ROSS

nected to the input of the receiver and then the gain of the signal generator turned up to a level of .1 volt or more. Then some local station, around 900 kc should be tuned in on the receiver and the signal generator dial rotated through the b-c band. You will note many beat note points on the dial.

Reasons for Beat Notes

The reason for the large number of beat notes can be traced to the three tuned circuits in the receiver, each one at a different frequency. These are the r-f, oscillator, and i-f stages. Innumerable spurious response combinations are possible. H. K. Morgan¹ has shown that the more common beats respond as follows:

- $nO \pm nS$, $nO \pm I/n$, nI + S/n, and nO + S/n
- where, O = oscillator frequency
 - I = intermediate frequency
 - S = signal or radio frequency
 - n =the harmonic order (i.e.,
 - first, second, third, etc., harmonic)

Where the complaint involves inter-

¹H. K. Morgan, Pro. IRE, p. 1,164; Oct. 1935

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ference of a second station interfering with the desired signal, the signal generator method can be used as a quick check for trouble. First the frequency of the beat notes should be compared with a listing of the local stations. For example, in the New York City area a whistle is often heard on WEAF, 660 kc. The interference has often been traced to a 1,560-kc signal, WQXR. Since the i-f frequency of many receivers is 455 kc, 910 kc represents its second harmonic. This, added to 660 kc equals 1,570 or 10 kc away from WQXR, causing a highpitched whistle on WEAF. In other words, WEAF and WQXR both come through the i-f when the receiver is tuned to 660 kc. For this condition, the i-f should be retrimmed to 460 kc.

Close-to-Station Locations

In locations where the receiver is close to a transmitter, a signal may be picked up at twice the station frequency. For example, a 650-kc signal may be received at 1300 on the dial. This is due to the second harmonic of the signal beating against the fundamental frequency of the local oscillator. The only possible cures are a well shielded chassis, or a wave trap tuned to the interfering signal, or both; Fig. 1. A large sheet of tinfoil or leadfoil cemented to the bottom of the cabinet will sometimes provide the necessary shielding. In this respect, the added shielding may detune the receiver due to the proximity of the shielding to unshielded coils mounted underneath the chassis. Therefore, when the set is aligned external to the cabinet, the chassis should be placed on a similar piece of tinfoil. In addition, glass-type converter tubes should either be shielded or replaced with their metal prototypes, making sure that the ground lug on the socket is grounded.

I-F Frequency Trouble

In some cases the i-f frequency is the cause of the trouble. For example, a signal of i-f frequency, usually an aircraft-beacon signal, may be the cause of the whistle. Or the second or third harmonic of the i-f frequency (for example for a 455 kc i-f, 910 kc is the second harmonic and 1365 kc the third) may cause the whistle.

When an i-f stage is energized with a signal, it becomes a source of radiating energy. This energy may be coupled back to the r-f section to cause the interference. Four methods may be used to prevent this condition. One involves the use of a wave trap in the



Fig. 2. Three methods of removing i-f interference. In *a*, a wave trap tuned to the i-f frequency is used in the antenna circuit. In *b*, an *RC* filter is used in the plate feed of the converter to remove any i-f feedback. A similar method for use in the detector circuit is shown in *c*.

antenna circuit tuned to the i-f frequency; Fig. 2a. The second method calls for the installation of a similar trap in the second detector. The third method requires the installation of metal shields under the chassis between r-f and i-f sections, or the rearrangement of the wiring so that pickup hetween r-f and i-f is minimized. In the fourth method an RC filter is used in the plate feed to the ri or converter stage; Fig. 2b. Sometimes, the installation of a simple RC network in the diode return is effective, since it prevents feedback through the avc system; Fig. 2c.

Cross-Talk

In older type receivers, a particularly troublesome type of interference is adjacent channel interference or cross-talk. This is usually evidenced by the presence of the interfering signal only when the desired signal is on the air. When the desired signal is off the air, the interference vanishes. The interfering station can be heard when the desired station signal is unmodulated. This may be remedied by reducing the coupling between primary and secondary in the antenna stage. It will be necessary to realign the antenna tuning capacitor trimmer after this is done. Where screen-grid tubes are used, the substitution of variablemu types may also eliminate the trouble.

Signal Rectification

External cross modulation is due to signal rectification in wires or other metal bodies in the vicinity of the receiver or its antenna. Poorly soldered connections, or poor contacts in the antenna, ground, power line, metal conduit, tin roof, or even metal window frames may be the source of the trouble. The trouble is usually evidenced as one or more signals being received at dial settings other than their assigned frequency. The new setting, where a single station is involved, will usually be some harmonic of the station. In two or more stations are the cause, the new setting will be any sum or difference of the station frequencies. This condition only arises where the signals are of high local intensity. This particular type of trouble is easily isolated by removing the receiver to the shop bench and noting if the condition still exists.

Remedies

If the interference is proved to be local, several remedies are possible. The antenna and ground should be checked for loose connections. In private homes where the BX cable is exposed, as in the basement, the ground clamp to the cold water pipe should be checked. Ground clamps should be added where possible. Where the station signal is being rectified in the power line, the trouble may show up as a strong modulation hum on the station carrier. A line filter or two .1-mfd capacitors across the line placed inside the receiver may remove this condition. For a-c/d-c receivers, a .1-mfd capacitor across the line helps sometimes.

Shielded Lead-Ins

If the trouble persists, a shielded leadin, or a low-impedance antenna (Continued on page 62)

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Fig. 1. Typical amplifier circuit using a double space-charge tetrode voltage amplifier of the hearing-aid type.

SUB-MINIATURES

by ROGER ETTON

THE wartime development of the tiny radio-proximity fuze receiver-transmitter³ using subminiature components and tubes has prompted the design of tiny pockettype receivers using sub-miniature parts and tubes. Engineering of the sub-miniature tubes has been quite advanced, several companies having already announced quite a few types.

These tubes are not to be confused with standard miniature types, which

(Continued on page 47)

¹Ralph G. Peters, *The Radio Proximity Fuze*, COMMUNICATIONS; October, 1945.

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Type	Class	Buib Size	Filamen Volts	t Rating	Use	Plate Volts	Negative Grid Volts	Screen Volts	Plate Current (ma)	Screen Current (ma)	Plate Resistance (ohms)	Ampli- fication Factor	Ohms Load for Stated Power Output	Undis- torted Power Output (milli- watts)
1C8	Triode Heptode	Diam. 3/8" Length 11/3"	1.25	.040	Converter	45	0	45	0.4	1.0	about 1 meg.			
1Q6	Diode Pentode	Diam. 3/8" Length 13/8"	1.25	.040	Det. Amp.	67.5	0	67.5	1.1	0.3	509,000	Gain=		
1V5	Pentode	Diam. 3/8" Length 13/8"	1.25	.040	Power Amp.	67.5	4.5	67.5	1.8	0.35	185,000	anout 50	25,000	50
1W5	Pentode	Diam. 3/8" Length 11/2"	1.25	.040	R-F Amp.	67.5	0	67.5	1.7	0.2	500,000		40,000	

Fig. 2 (above). Tentative characteristics of Sylvania sub-miniature proximity luze style tubes.

		Fila	ment	Plate	Menotivo	S	Blate Current	Second Contract	Plate
Tube	Туре	Volts	Amp.	Volts	Grid Volts	Volts	(ma)	(ma)	(ohms)
1W5	sub-miniature	1.25	.040	67.5	0	67.5	1.7	.4	500.000
2E32	sub-miniature	1.25	.050	22.5	0	22.5	.35	.3	350,000
10724				45	0	45	1.7	.7	350,000
114	miniature	1.4	.050	90	0	67.5	3.5	1.4	500,000
1N5	standard	1.4	.050	90	0	90	1.2	.3	1,500,000

Fig. 3 (above). Comparison of sub-miniature, miniature and standard tube characteristics.

Туре	Filament Voltage	Filament Current	Maximum Heptode Plate Voltage	Maximum Heptode Sciten Voltage	Maximum Triode Plate Voltage	Maximum Total Cathode Current	Minimum External Signal Grid Bias	Frequency Converter	
2G22	1.25	50 ma	45	45	45	2.0 ma	0 volts	Plate voltage (heptode) Screen voltage (heptode grids 2 and 4) Oscillator plate voltage (triode) Signal grid bias (heptode grid 3)* Oscillator grid resistor (triode) Plate current (heptode) Oscillator plate current (triode) Oscillator grid current (triode) Conversion plate resistance (approx.) [*] Grid resistance = 5 megohms	22.5 22.5 22.5 50,000 olimis 200 µa 300 µa 1 ma 30 µa 0.5 meg

Figs. 4 (above) and 5 (below). Fig. 4 shows the preliminary characteristics of a Raytheon triode-heptode sub-miniature frequency converter. In Fig. 5 are preliminary data on the Raytheon diode-pentode sub-miniature detector-amplifier.

Туре	Filament Voltage	Filament Current	Maximum Plate Voltage	Maximum Screen Voltage	Maximum Cathode Current	Operating Characteristics	
2E42	1.25	30 ma	45	45	1.0 ma	Plate voltage Screen voltage Control grid voltage* Plate current Plate resistance = 5 megohms	22.5 22.5 0 volt 0.4 ma 0.15 ma 0.25 meg
						RESISTANCE COUPLED AMPLIFIER	
						Plate supply voltage Screen supply voltage Load resistance Screen resistance Voltage gain	22.5 22.5 1 meg 5 meg 20
						DIODE RATINGS	
						Minimum diode current with 10 volts d-c applied Maximum diode current for continuous operation (The diode plate is located at the negative end of the fi	0.5 ma 0.25 ma ilament.)

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A Ν

by ROBERT L. MARTIN

Fig. 1. Diode detector circuit used in G.E.L. 651. Diodes are con-nected in parallel for minimum diode resistance.





Fig. 2. Battery model receiver, Ward 14 BR-474, using a combina-tion detector-avc-am-plifier circuit. Note the small audio coup-ling capacitors used to minimize cabinet reso-nance. The r-f bypass capacitor at the first audio output usually has a value of 50 or

100 mmfd.

Fig. 3. Detrola 408 battery - phonograph model with a combination detector system.

NE of the most interesting sections of the modern receiver is the avc/detector/a-f combined circuit, featuring the diode detector. Diode detectors are preferred in these circuits because their greater linearity permits less distortion and because of the simple means of obtaining avc from the detector. These advantages far outweigh the disadvantages, such as reduction of selectivity due to the current drawn by the detector and the absence of amplification in the detector.

Diode Detector Tubes

Almost any tube at all will serve well when connected as a diode detector for intermediate frequencies. The 6H6 dual diode was originally introduced as a detector, but has been largely supplanted by the more economical diode-triode, duplex diodetriode and diode-pentode combination types of tubes. The triode sections are available in both low and high mu, the low-mu tubes originally being used with audio coupling transformers. Typical tubes used included 6R7, then the 6SR7, 6ST7, etc. The high-mu triodes and pentodes are designed for resistance coupling, providing an audio gain of 20 to 50, the latter figure being common for pentodes.

Gain is the principal objective in a loop-fed portable receiver because of the limited pickup, not only in the first audio amplifier stage but throughout the set. In other words, quality is exchanged for gain. In large sets, on the other hand, it is usually permissible to use another tube or two. Thus the sacrifice is in the other direction, gain for quality.

G. E. L-651

One of the simplest of diode detector circuits, in a duplex-diode-triode application, is shown in Fig. 1; G. E.

*See SERVICE, August, 1945, for additional data on second detector-avc systems.

22 • SERVICE, DECEMBER, 1945

DETECTOR SYSTEMS*



L-651. The diodes are connected in parallel for minimum diode resistance in a half-wave rectifier circuit. The diode-load resistor is the volume control itself, the usual values being $\frac{1}{2}$ to 2 megohms. Battery sets generally use higher values than standard a-c or a-c/d-c receivers. The bypass capacitor shunting the diode-load resistor is necessary for detection and also serves to bypass the unwanted i-f to ground. Typical values for this capacitor run from 100 to 250 mmfd. Too low a value leads to insufficient bypassing which often results in screetching signals, while too high a value will bypass the high audio frequencies, causing a loss in treble.

Audio Capacitor Values

The a-f coupling capacitor between the volume control and the first a-f grid usually lies between .001 mfd and .01 mfd depending upon the value of the grid leak and the proportion of bass response desired. For instance, a .001-mfd or .002-mfd capacitor will attenuate the lows, minimizing hum output and giving a greater apparent power output. This is important in the less expensive receivers and portables. Since the ear is much more sensitive to the middle range than to the bass, a given amount of power will appear increased when the bass is removed. Most first audio stages are biased by means of grid leaks with values of 2 to 15 megohms, the higher values being used in battery sets. The product of the coupling capacitor and grid leak values must not be excessive or the time constant will be to great. Then the change in volume will lag behind the manipulating of the volume control. The optimum values for capacitor and leak for the average receiver are .005-mfd and 6.8 meg-The circuit shows a shunt ohms. capacitor at the 35L6 input used for tone control,

Ward 14-BR-474

A detector/avc/audio amplifier circuit similar to that shown in Fig. 1, used in this instance in a battery Fig. 4. RCA QU52C with a single diode detector and bass-compensated volume control. A .005-mfd capacitor and 10-megohm resistor are used for audio coupling.

Fig. 5. The G.E. 20 receiver using an extra amplifier stage to provide for the loss in gain caused by degeneration. A tertiary winding on the output transformer supplies the feedback voltage which is fed to the cathode of the 6J5 through a 3300-ohm bias resistor.

Fig. 6. Ward's 62-271, with a triode used as a diode; grid is plate and plate is grounded. Enforced cathode bias is applied to the first audio instead of the usual cathode resistor. This is done by bleeding some d-c from the B supply through a potentiometer.







model is shown in Fig. 2; Ward 14-BR-474. Small first audio coupling capacitors have an additional advantage in portables, minimizing cabinet resonance. This is particularly important in portables because the case is all enclosed. The triode plate resistor in battery sets usually has a value of 1/2 or 1 megohms in contrast to $\frac{1}{4}$ to $\frac{1}{2}$ megohms for larger receivers. The higher values permit an increase in gain, necessary in portables. In addition with higher values of resistors, smaller coupling capacitors are used. An r-f bypass capacitor of 50 or 100 mmfd is usually employed at the first audio output.

Detrola 408

A Detrola 408 battery-phonograph

Figs. 7 (above, left) and 8 (above, right). In Fig. 7 appears the combination detector system of a 6-volt 150-mil farm receiver, RCA QB1. A series resistor precedes the volume control for r-f filtering. In Fig. 8 we have the 6B7 double-diode pentode as a detector-amplifier; Radio Wire Television B92. A 41 follows this tube as a triode driver.

combination circuit is shown in Fig. 3. The input to the volume control is switched from detector output to the

•

Figs. 9 (below, right) and 10 (below, left). The National NC-44B communications model with a 6SQ7 detector amplifier system appears in Fig. 8. Cathode bias is used here instead of the usual grid-leak bias. Fig. 9 shows the 3A8GT combination detector system of the Motorola 41H. The single diode detector is fed by the pentode section serving as an i-f amplifier. phono pickup. A .001-mfd capacitor is used. A pair of 3Q5 output tubes are connected in parallel.

RCA QU52C

A 6SQ7 single-diode detector and bass compensated volume control are featured in the RCA QU52C, shown in Fig. 4. A 22,000-ohm resistor is conected in series with the ½-megohm control. The latter is tapped at 125,-000 ohms for the compensating shunt consisting of a .025-mfd capacitor and 12,000-ohm resistor. This receiver uses a .005-mfd capacitor and a 10megohm resistor for audio coupling. An iron core i-f transformer feeds the detector diode. The other diode is grounded. A variable tone control,

(Continued on page 49)







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VOLUME AND TONE

THE variable wire-wound and carbon-element resistors employed as volume and tone controls are far more subject to trouble than are *fixed* resistors. This is so because their vital parts are subjected to handling, mechanical motion and some wear almost every time the set owner turns the receiver on or off or tunes to a new station and subsequently adjusts the volume or tone to the desired value. Consequently, they receive even more operation cycles than do the tuning dial and capacitor.

This repeated operation of the contact arm eventually causes wear of the moving parts and of the resistance element. The latter, because of the nature of the resistance materials that must be employed¹, is not particularly rugged even when new.

Locating Difficulties

As a rule, faults that have developed in volume and tone controls quickly make their presence known, because rotation of the control knob of the unit while the receiver is in operation results in annoying noise, intermittent operation, or otherwise unsatisfactory control of the volume or tone. Such performance is extremely annoving to the set owner-especially in the case of a volume control-so he usually does something about having it repaired. As a rule, volume and tonecontrol troubles are responsible for a substantial percentage of the Service Man's repair and replacement work.

Common Troubles

(1)—Change in Resistance: Wearing away of the carbonaceous resistance film, or the fine nichrome wire, employed for the resistance element¹ will cause its cross-sectional area to decrease, and the total resistance of the control to increase. However, this resistance increase in itself is not ordinarily of great importance, because exact values of maximum resistance are fortunately not required in ordinary volume or tone-control circuits. Variations of as much as 20% or even more in the total resistance of such controls are perfectly satisfactory in ordinary receiver circuits. Only where the volume control is employed as a part of the power voltage-dividing system is it necessary to observe fairly close tolerance.

However, if the control unit em-

1Part 6, SERVICE, August, 1945.

Part Ten of a Series on Receiver Components

by ALFRED A. GHIRARDI Advisory Editor

ploys a critical resistance taper, such wear may alter it and the smoothness of the control action. Also, if the unit is employed as a current-carrying type of control, excessive wear may cause overheating and eventual distruction of the resistance element, as will be explained later.

(2)-Noise: One of the most important demands made upon a volume or tone control is that it be electrically noiseless in operation. The chief objectionable result of continued use and wear of the resistance element is the development of noise when the control is manipulated. This is caused by sudden variations in resistance that result as the contact arm is moved. These sudden resistance variations cause corresponding variations in the currents or voltages controlled by the volume or tone-control circuit, resulting in sudden current or voltage impulses. These, when amplified enormously by the high-gain receiver circuits, are heard in the loudspeaker as loud, scraping noises.

Noise, or even intermittent operation, usually results from the fundamental cause of *imperfect contact*. If imperfect or variable contact develops between the resistance element and the contact-making element¹ at any point in its travel, or at any other point in the electrical circuit of the control, noise or intermittent operation

This series of articles on Fixed and Variable Resistors has been especially prepared by Mr. Ghirardi in response to the numerous demands of newcomers to servicing for detailed construction, operating and servicing information on the various components that go to make up modern receivers.

This series constitutes a valuable reference which even experienced Service Men will find worthwhile as a refresher and for new servicing ideas. We would appreciate receiving suggestions for the subjects of future articles in the series.—Ed. will result. This may be brought about by any of several conditions, the most common of which are: (1)— Loss of spring tension (or non-uniform spring tension) in the contact arm; (2)—dirt (or the formation of a glazed non-conducting surface) on the resistance element, the contact surface, or the shaft slip-ring and wiper contact surface; (3)—loose or corroded terminal rivets.

Manipulation of the volume or tone control shaft while the receiver is turned on will reveal this trouble either by the *scratchy* noise issuing from the loudspeaker or failure of the unit to smoothly control the volume or tone.

Volume controls used in diode circuits are often unnecessarily replaced because of apparent noisy operation when, in many cases, the noise is due to excessive diode current flowing through the volume control. It is good practice to test or replace the tube associated with the diode control, to assure that any abnormal diode current condition does not exist, before changing the volume control. Changing the volume-control circuit to the more recent type in which no diode d-c flows through the volume control will cure noise due to this cause in all diode controls. The circuit change recommended will be presented in the next article of this series.

Open-Circuits

(3)—Opens: In both carbon and wire-wound controls (especially the older types), movement of the contact arm directly over the surface of the resistance element may eventually wear away the carbon deposit or the nichrome resistance wire sufficiently to create an open circuit. However, opens caused directly by wear are not the most common. A control will become excessively noisy long before this condition is reached. Opens are more frequently the result of a secondary action. If the resistance element experiences much wear, its resistance and the heat generated in it in the worn area increase while its heat-dissipating capability decreases. Consequently, if it is being employed as a current-carrying type of control, such wear may cause it to become overloaded by normal current, or momentary excessive current. This may eventually cause an intermittent or full open by development of hot-spots and burning

CONTROL RESISTORS

out of the resistance element. An ohmmeter check will reveal this condition.

An *open* volume control will not provide proper control of volume, even though it may permit partial transfer of the signal in some volume control circuits.

(4)—*Shorts*: Because of the way volume and tone controls are designed, *shorts* do not often occur in them. However, in those units in which the metal case is *hot* and an insulating bushing is employed between the chassis and the mounting bushing of the control, a defective insulating bushing or washer will often create a short to the chassis. Such trouble can usually be located only after all leads to the control have been unsoldered and an ohmmeter test is then made between each terminal of the control and the chassis.

Repair Versus Replacement

There is no doubt but that the best service policy is to *replace* a faulty volume or tone control with a suitable replacement unit whenever this is possible, for a repaired unit seldom gives as long or satisfactory service as does a new one.

There are general recommended procedures for selecting the correct replacement volume and tone controls for standard listed receivers, orphan and unlisted receivers. In many instances it is more expedient to repair than to replace such controls. The methods of repairing all types of volume and tone controls will be explained in the next article of this series.

"Standard" Listed Receiver Replacements

(1)—Use of Manufacturer's Replacement Lists and Guides: Sufficient accurate manufacturer's volume and tone-control replacement information is now available to Serivce Men to permit the selection of the correct replacement unit for almost all standard listed receivers (those manufactured by the standard name receiver manufacturers). Most receiver manufacturer's service sheets supply specifications for the proper control to use. In addition, schematic diagrams of

manufacturer	tian	Control	Swill	ch Shaft	12	Ohms
	0.00	DIC		BUC	NOOD	ADM
EMERSON	R/	ADIO	ð.	РНО	NUGR	APH
375LW	Vol.	+013-13	7	.41	16	1Meg.
409, 410, 411 (&4 Ch)	Vol.	•015-11	9	.41	16	20M
409, 410, 411						
(U4C Ch.) 415, 416 (V4LA	Vol.	014-12	25	.41	36	75M
Ch.)	Vol.	*D16-11	9	.41	10. 16	20M.
420 (V4 Ch.)	Vol.	+D16-11	9	.41	10. 16	20M .
867	Vol.	J-525.				500M
678; (2nd type)	VOI.	J-938.	tram	original h	867	300111
	NS	Obtain	from	original h	Afr	
1 753	Vol.	01-50		.s	4, 8, 10	5M
L. / J.J	Tone	D11-12			16	25M .
M758	Vol.	D J-50		. s	4, 8, 10	5M
	Tone	D11-12	20		16	25M
\$755	Vol.	DJ-50		.s	4, 8, 10	5M
770 (AW7 Ch.)	Vol.	*D13-13	37	.41	16	IMeg
	Ton	D11-12	20		16	25M
965	Vol.	J-525		.5	10	IMeg
CS	Vol.	016-11	15	.41	16	10M
E. KS	Vol.	P16-1	10	.41	16	15M
10	VOI.	016-1	18	41	16	10M.
13	Tool	*D13-12	73		16	50M
KS	Vol.	*D14-1	16	.41	16	10M
T, TS	Vol.	D16-1	16	.41	16	10M
	LE	CTRI	CAL	PRO	DUCT	rs
an sw	Vol	+013-1	33	.41	16	500M
51	Vol.	+013-1	33		16	500M
71	Vol.	*013-1	33	.41	16	500M .
400AC	Vol.	+014-1	16	. 41	16	10M
40#DC	Vol.	D16-1	16	.41	16	10M.
458A	Vol.	•D13-1	33		16	500M.
4608	Vol.	+013-1	33		16	500M
479C	Vol.	•013-1	33		16	500M
4800	Vol.	*D13-1	16	41	18	10M
300AG	Too	*012.1	21		16	50M
500000	Vol	D18-1	16	. 41	16	10M
550AC	Vol	016-1	16	41	16	10M
	Ton	e*D13-1	23		18	50M
600AC	Vol.	+D14-1	16		16	10M
	Ton	e • D13-1	23		16	50M
700 D C	. Vol.	D16-1	16	41	16	10M
	Ton	e •013-1	Z3		16	DUNN
ERLA (Se	e I	Electr	ica	Rese	arch	Labs
ESPEY M	FG.	ço.	IN	C.		
875	Vol.	*013-1	33	41	16	500M.
	100	4D12 1	33	41	16	SOOM -
5111	Vol	*D13-1	133		16	SOOM
7111	Vel	*D13-1	33		16	1Men
////	Tor	w + D13-1	33		16	500M
P181 7163	Vol	*D13-1	137	41	16	1 Med
(101. (103						



the typical volume and tone-control circuits² employed in receivers, and the recommended type of control to be employed in each one; comprehensive lists, manuals and guides that



specify the type number of the particular manufacturer's replacement control recommended to be used in each receiver model listed (see Fig. 1); and many other helps are distributed to Service Men by leading manufacturers of replacement controls. Every Service Man should collect as many of these control manufacturer's manuals and guides as he possibly can, and make a practice of consulting them in his control replacement work. A surprisingly large amount of useful general information about the various types of controls can also be picked up from them.

Exact Duplicates and Standards

(2)-"Exact Duplicates" and "Standard" Units: When consulting the replacement lists and guides of some control manufacturers, it will be noticed that in many instances two or more types of controls are listed. One type is the exact duplicate, mechanically and electrically, of the original control in the receiver and is guaranteed to fit and operate without mechanical or electrical changes. These controls have shafts of the proper type and length, taps and internal C -bias resistors when required, proper switches mounted, and incorporate any special mechanical or electrical features required.

Inasmuch as it would require a considerable amount of time, trouble and expense to work up a complete line of individual *exact duplicate* replacement controls for use in the several thousand models of standard receivers now in use, many listings are mainly suggestions for using *standard model* controls units closely approximating the requirements of the receiver so that they can be satisfactorily installed if certain minor mechanical changes (if any) such as alteration of shaft length, bushing, etc., are first made in them.

List of Checks

The Service Man will do well to distinguish between such listings of *exact duplicates* and *standard* units. The former are in accordance with the original specifications of the set designer and manufacturer; the latter

Fig. 2 (left). Customary terminal designation for volume or tone controls when facing shaft end of control.

²For a comprehensive explanation of the various volume and tone control circuits employed in radio receivers see *Volume Control Circuits* by Robert L. Martin, SERVICE; June and July, 1945.

are simply improvisions that have been found acceptable in practice. The *exact duplicate* type should be employed as a replacement whenever it is available.

Unlisted, Orphan and Private-Brand Receiver Replacements

Frequently, orphan, private-brand, loft or other such receivers are encountered for which no published volume or tone control replacement information or schematic diagram is available. In such cases, the matter of selecting the correct replacement control becomes somewhat more complicated. The following procedure is recommended³:

(1) Before removing the faulty control, or even unsoldering the leads, the control circuit should be traced and then a schematic circuit diagram should be made. Also, it is important to note the connections of the leads to the control and make a pictorial sketch of the lead connections to prevent wiring mistakes when connecting the leads to the new control later.

(2) The faulty control should be removed from the chassis.

(3) Then the diameter and depth of the control should be measured and recorded.

(4) The shaft diameter and length can then be measured and recorded and a note made of the type of end provided for the knob.

(5) The next step involves the measurement of the overall resistance of the control with an ohmmeter. If the resistance element is open, it will be necessary to open the control and measure the value of resistance to each side of the open. The overall resistance can then be arrived at by adding these two values. If the element is burned out, it will be necessary to measure the various remaining sections of the resistance, and from their total, estimate the amount of resistance that has been destroyed, so as to obtain the original overall resistance value.

(6) We then determine the maximum wattage the control will be called upon to dissipate in normal use. The maximum current it will called upon to carry can be determined by inserting a d-c milliammeter in series with its circuit, with the receiver turned on. If its resistance element is *open*, a fixed resistor of equal value may be substituted temporarily. The maximum wattage the resistor must dissipate⁴ may then be calculated by



Fig. 3. Rear view of wire-wound control having an "open" resistive element. The terminal designations L and R indicate left- and righthand positions. Method of checking the taper in a control of this type is explained on page 30.

multiplying the square of the maximum current in *amperes* by the resistance of the control in ohms, i.e., I^2xR . This quantity is especially important in the replacement of carbon-element type controls in bias-type control circuits in cases where two or more stages are controlled through the unit, or where considerable bleed current is carried.

(7) It is now necessary to note whether the control is of the wirewound or the carbon type. Wirewound controls are generally used for values up to approximately 10.000 olims, and usually for antenna-shunt control, bias control (or a combination of both) and for voltage-dividing applications. This is not always the case, however, for carbon-element controls are made in values of maximum resistance ranging approximately from 1 to 20,000 ohms (and in 2-watt and 4-watt ratings). Carbon-element controls are ordinarily used in applications where a total resistance above approximately 10,000 ohms is required.

(8) Since the taper or resistancevariation characteristic of a volume

⁵The various resistance tapers obtainable in standard replacement controls, and the type of control circuit each is particularly suited for, were described and illustrated in *Part 7*, SER-VICE; September, 1945. or tone control actually is more important than its exact overall resistance, the taper of the faulty control should be ascertained in order to select a replacement that will duplicate the action obtained with the original control.

One way of determining the type of taper required is to study the volume control circuit employed, as determined in (1). When this is known, the proper type of taper required for proper operation in this type of control circuit may be ascertained by reference to the replacement control manufacturer's catalog⁵. This method is especially useful in cases where the resistance element of the control has been so badly damaged that it is difficult (or impossible) to make resistance measurements on it.

The original taper can be determined with fair accuracy by measurement if the resistance element (either wire or carbon) is in fair mechanical condition, even though it may be *open* in one or more places.

Determining Taper

To determine the taper of a control wherein the resistance element is *not* open, that is, where there is continuity between the left hand terminal, L, and the right hand terminal, R, the following procedure should be applied:

(1)—Place the control in the position shown in Fig. 2, with the moving arm and shaft set at the center of its rotation.

(2)—With a suitable ohmmeter measure the resistance between terminal L and terminal C. Make note of the value.

(3)—Measure the resistance between terminal *C* and terminal *R*, and compare the last reading with the first.

If the resistance between terminals L and C is *lower* than the resistance between terminals C and R the control has a left-hand taper.

On the other hand, if the resistance between terminals L and C is *higher* than that between terminals C and R the control has a righthand taper.

If the resistance of the two halves of the control is the same on both sides of the center of rotation of the moving arm, i.e., when the resistance between terminals L and C is of the same value as that between terminals C and R, the control is without taper and is therefore called a linear control because the resist-

www.americanradiohistorv.com

³Courtesy Yaxley Replacement Volume Control Manual, P. R. Mallory & Co., Inc.

⁴It should be remembered that the maximum safe power-dissipation rating (applying to the maximum-resistance setting) of *tapered* controls is *less* than that of corresponding linear controls of similar physical size, depending on the resistance-taper curve. In general, those controls having the steeper taper curve have the lower rating.

rating. Also, the wattage rating specified by the control manufacturer always applies to the maximum-resistance setting only. For intermediate points of rotation, the dissipation rating in watts decreases—the amount depending upon the steepness of the taper of the resistance element. For a detailed discussion, see Part 7, SERVICE; September, 1945.

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ance increases linearily with rotation of the moving arm.

To determine the taper of a control wherein there is an open in the resistance element, the following procedure can be applied:

Fig. 3 should be referred to during this check. This diagram shows a *rear* view of a wire-wound control with an *open* at the point marked X. It will be noted that the terminals bear the designations left-hand, L, and right-hand, R, as they are when facing the *shaft end* of the control.

(Although a wire-wound control is shown, these instructions also apply to carbon type controls.)

(1)—To determine the taper, the moving arm should be placed in the center of its rotation, as in Fig. 3.

(2)—The resistance between terminal R and terminal C should be measured and a note made of value.

(3)—Then the resistance between C and the edge of the open marked B should be measured.

(4)—We now have to measure the resistance between terminal Land the edge of the *open* marked D.

(5)—In the last step we have to add the values obtained in steps three and four, to obtain the resistance of right-hand half of control

With the values of the two halves of the control known, a comparison will quickly indicate what the taper is, as was explained previously.

Checks for Additional Opens

If there is more than one open, the above method should be followed, with the exception that the values of resistance between the different open places will have to be measured and then all added together to compare resistance of the two halves of the control.

The foregoing method of determining the taper by comparing the resistances of the left and righthand halves of the resistance element is a rough and ready method applicable in most cases. However, whenever it is necessary to check the *exact shape* of the taper characteristic employed in any control, it



Fig. 4. Method of checking the resistance taper of a volume or tone control. At a appears the method of connecting the cortrol and nhmmeter, while at b we have a plot of the readings taken.

may be done very readily by mounting the control on the rest of the panel and fastening it in place with its terminals pointing downward, as shown at (a) in Fig. 4. A knob (preferably one of the bar, or pointer, type) should be placed on the shaft and adjusted so that when the shaft is turned all the way to the left (just before the on-off switch snaps off) the position of the knob pointer can be marked in pencil on the panel. The knob should be fastened in this position, and position of its pointer marked. Then the knob should be turned all the way clockwise and the position of its pointer marked again. This total rotation should then be divided into 10 equal parts, by means of radial lines drawn on the panel.

An ohmmeter is then connected to the left-hand terminal L and to the center terminal C of the control. Now, with the control rotated all the way to the left (just before the on-off switch snaps off) the first resistance reading is taken, which in most instances will be zero. We continue by taking a reading of the resistance at every 1/10 of the total rotation, from left to right. Then the readings are plotted on graph paper, as illustrated at (b) of Fig. 4. The graph produced will indicate the taper of the control. That illustrated at (b) indicates a typical left-hand taper.

(9) If the control is tapped, the resistance between the left-hand terminal L, and the tap should be measured.

(10) At this point the following information will have been collected:

(a)-Control circuit employed.

- (b)—Diameter and depth of control.
 (c)—Shaft dial diameter, length, shape.
- (d)-Overall resistance.
- (e)—Maximum wattage the control is called upon to dissipate.
- (f)—Type (wire-wound or carbon). (g)—Taper.

(h)—Tap value (if control is tapped). All of this information is neces-

sary for intelligent selection of the correct replacement control from the control manufacturer's catalog.

Useful Suggestions about Replacement Controls

The physical size of the replacement control may, or may not, have to be the same as that of the original unit, depending upon the available space. Replacement unit must be small enough to fit into the space.

The shaft-length diameter and type of end should be such that the unit works satisfactorily when installed. The use of the newer types of midget controls designed to be used with plug-in shafts makes it possible to replace, both mechanically and electrically, over 90% of the midget control requirements by carrying a small stock of only five or six controls and a kit of the various types and sizes of plug-in shafts.⁶

Carbon-element controls which have abrupt tapers generally cause *noise* troubles sooner than those having more gradual tapers. Practically all volume controls designed for audiocircuit use have a taper that is gradual over approximately the first 50% of rotation and much steeper thereafter, but some of these units have a much more abrupt change of taper in the 50-70% rotation region than do

(Continued on page 51)

MOUNTING BUSHING

COUPLER SHAFT

"Part 9, SERVICE, November, 1945.

Fig. 5. High voltage insulating coupler applied to control that must be used in high-voltage circuits as in television and e-r oscillographs. Sectional view of coupler construction appears below. Further data on coupler and use appears on page 51. (Courtesy Clarostat Mfg. Co., Inc.)



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Stromberg-Carlson industrial designer's Below conception of a television sound receiver of the future.





Above and below, G.E. a-c/d-c tube superhets. Model above is for broadcast only, while receiver below covers broadcast and short wave bands.



Below, Stewart Warner 7-tube 3-band a-c model with iron-core push-button tuning for broadcast short-wave bands. Audio system uses inand verse feedback.



Below, Stewart Warner 6-tube a-c broadcast band Output is 4 wat impedance loop. watts. combination. Has a low



Below, Belmont 6-tube a-c/d-c broadcast band model, using 12 SK7s as r-f and i-f amplifiers, 12SA7 as a converter, 12SQ7 detector-audio, 35L6GT power output and 35Z5GT rectifier.



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ITH high-fidelity receiver popularity growing daily, many Service Men are showing quite a lively interest in these models. As a result SERVICE has been receiving requests for analyses of high fidelity types, such as the Ward 10-tube chassis, shown in Fig. 1; model 93WG-1000/1001. This is an interesting receiver with several unusual features.

A 6SK7 tuned r-f stage operates

by HENRY HOWARD

from loop or foil plus external antenna. With antenna operation on the broadcast band, a supplementary 5-mmfd

Fig. 1. The Ward 10-tube high-fidelity receiver.

capacitor provides capacity coupling for increasing the high-frequency sensitivity. A multi-turn broadcast primary is bypassed by a 250-mmfd capacitor to allow short-wave signals to get by without attenuation. The bandswitch is arranged to short lowerfrequency coils in high-frequency operation in all three sets of tuned circuits

(Continued on page 42)



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SERVICING THE OSCILLOGRAPH

T HE effectiveness of external synchronization may be tested through the simple expedient of tracing the waveform with a fountain pen, or with a china marking pencil, subsequently reversing the a-c voltage applied to the input terminals of the vertical amplifier. It will be observed that the new trace written by the beam is exactly 180° out of phase with the pattern traced on the screen of the tube under the conditions described in last month's discussion.

If the synchronization control switch is now turned to the INT position, and the synchronization control potentiometer is advanced slightly to secure good synchronization, it will be observed that the new pattern written on the screen is exactly 90° out of phase with the pattern previously traced. Further, it will be found that reversal of the a-c input voltage which is impressed across the vertical amplifier input terminals effects a 180° reversal of the wave pattern, but that the pattern is still 90° out of phase with the trace marked on the screen. Obviously, under one of the preceding conditions, the new wave pattern is 90° out of phase with the marked pattern, and under the remaining condition, the new pattern attains a 270° phase relation with respect to the marked pattern. Finally, if an a-c input potential which is readily varied in phase is substituted for the a-c input potential, it will be observed that variation of the relative phase position of the input voltage effects no alteration in the phase position of the pattern, whereas, if the synchronization control switch is again operated to the EXT position, the variation in the phase position of the input voltage is readily reproduced as an appropriate motion in the pattern traced.

Sine-Wave Signal Voltages

Under certain conditions, it is necessary to effect other deflection of the beam along the horizontal or X axis, than that provided by the linear timebase oscillator system. Thus, it may be desirable to apply a sine-wave signal voltage to the input terminals of the horizontal amplifier. Here, it is necessary to reduce the intensity of the [Part Six of a Series]

by S. J. MURCEK

electron beam to zero to prevent possible damage to the electron screen. The linear time-base oscillator coarsefrequency control is then rotated to the off position, which connects the input terminals H and G of the horizontal amplifier to the actual input circuit of the horizontal amplifier. The desired voltage source may be then connected to these terminals, and the intensity of the beam advanced carefully until a horizontal trace appears on the screen. If no trace appears, the luminous spot being clearly visible as a spot, the intensity control must be returned to the zero position, and the horizontal gain control advanced. Then a second test may be made to ascertain if horizontal deflection of the beam is present. Once a horizontal deflection of the electron beam is obtained, the brightness of the pattern may be increased to the desired level by clockwise rotation of the intensity control.

Intensifier Electrode

Where an intensifier electrode is used in the oscillograph tube, a suitable intensifier voltage control switch is usually mounted on the control panel. This switch must be left in the off position until a suitable recurrent pattern is written on the screen. Then the switch can be thrown to the on position to intensify the luminosity of the written wave pattern. In general, however, it will be found that the use of the intensifier electrode is usually not advantageous, except under the condition that voltage impulses of short duration and high voltage level are under observation, or under the condition that the pattern written is to be photographed. In the former, the intensifier electrode serves to render visible extremely sharp wave fronts or cusps which are not otherwise visible to the eye, and in the latter, the increased clarity and heightened luminosity of the wave pattern is necessary if the image is to be properly recorded on photographic film.

Television-Type Tubes

Oscillographs incorporating television-type tubes which utilize an intensifier electrode generally incorporate a separate rectifier and filter power supply system for the development of high voltages required. The intensifier electrode voltage supply actually consists of the voltage available from a low-voltage source, and from a second power supply which rectifies and filters the total potential.

Protection of the oscillograph against mechanical vibration and severe physical shock is of prime importance. The instrument should never be placed on the housing of a motor or transformer which vibrates or hums audibly. Where it is necessary to place the instrument near a forging hammer, or near the source of sharp, loud sonic impulses, such as those arising from the discharge of explosives, the cabinet of the device should be placed on a sponge rubber mat to protect it from severe physical shocks. These precautions are necessare to protect the electrode alignment and the heater tube structure as well as those of the remaining tubes in the instrument.

Magnetic Leakages

The magnetic leakage field present in the immediate vicinity of transformers and motors, as also in the welding throat of resistance welding devices, is also of considerable importance. Therefore, the oscillograph must be placed well away from the sources of such leakage fields since the penetration of the field through the magnetically permeable cabinet of the device is unavoidable. Where a strong leakage field is present within the cabinet of the oscillograph, the image written on the screen will be distorted, giving a false representation of voltage variations under observation. Further, if the field is of considerable intensity, the heaters of the picture tube and the supporting tubes are subject to magnetronic motoring. Under the latter condition, if the leakage field is a static



field, the a-c field built up around the tube heaters by the heater currents reacts with the stray leakage field to provide action motion. This is in the form of severe vibration to the tube heaters,' subjecting these to failure from crystallization of their metal structures. The heater of a tube which operates in a strong magnetic leakage field becomes extremely brittle and is subject to early fracture with slight mechanical shock.

Since the screen is exposed to damage or breakage by flying foreign objects, the cabinet must always be placed so that this hazard is reduced to a minimum. The hazard to the tube is reduced considerably if a calibrated transparent plastics chart is placed before the screen.

Mechanical Wear Problems

Despite the maximum amount of protection and care, mechanical wear of certain components or eventual failure of capacitor dielectrics or of certain types of resistance elements may cause instrument failure. In genFig. 1. Circuit of Du Mont 164 E oscillograph.

eral, cessation of oscillograph operation occurs with the open-circuiting of a potentiometer resistance element, failure of an operating switch contact, or puncture of the dielectric in a capacitator due to its deterioration.

Component failure may interfere with satisfactory vertical deflection. This trouble becomes evident when the luminous spot does not write the vertical component of a test screen pattern. The defect may be due to (1)— Burnout of the vertical amplifier pentode heater; (2)—open circuit in the extreme maximum end of the vertical gain-control potentiometer; (3)—puncture of the pentode screen grid capacitator; (4)—open-circuited plate coupling capacitator; or (5) open pentode plate-loading resistor.

Rapid Tests

Rapid diagnosis of the actual defect is facilitated in the complete analysis may be accomplished without removing the oscillograph from the cabinet. In testing the defective vertical gain potentiometer, for instance, the shaft of the potentiometer may be first turned to the extreme maximum position in order that the slider arm contacts the ungrounded positive terminal of the resistance element. Then application of a small signal potential, at this juncture, should effect vertical deflection of the beam. If, however, no vertical deflection of the beam occurs, it is apparent that the vertical gam potentiometer resistance element is intact.

The succeeding step in the analysis of the vertical deflection system defect involves the removal of the vertical deflection-plate connecting link at the rear of the cabinet. With this link removed, an a-c potential of approximately 100 volts must be impressed between terminal H and GND. If the linear-time base system is properly adjusted, it should then be possible to secure vertical deflection of the beam. Here, it should be observed that failure of the a-c test potential to deflect the



beam along the vertical axis indicates the existence of a defect within the kinescope tube or its socket. These conditions are not generally encountered in normal service, unless the oscillograph has been subjected to high vertical deflection potentials or has been used in a damp atmosphere.

If it is found necessary to remove the oscillograph from the cabinet the high-voltage circuit rectifier tube should be removed from its socket. This will permit operation of the vertical amplifier low-voltage circuits, and the hazard of exposed high-voltage circuits will be minimized. With the removal of the high-voltage rectifier diode, the oscillograph may be connected to an a-c power source for further circuit examination.

Localization of Defects

Continued procedure in localization of the vertical amplifier system defect begins with the removal and testing of the vertical amplifier pentode. Failure of the latter is a common defect encountered in vertical deflection systems. Should the pentode prove to be satisfactory, it is then necessary to take a voltage reading with a voltmeter having an internal resistance of 1000 ohms-per-volt. This check should be made between the plate contact in the pentode socket and the oscillograph chassis. The reading should be about 125 volts. This voltage also indicates that the pentode plate-loading resistor is intact, necessitating further testing of the amplifier circuits. A voltage reading taken between the screen-grid contact in the tube socket and the chassis of the oscillograph must range between 85 and 90. A reduction from this value, or its complete absence indicates that the screen electrode bypass capacitator C₃ in Fig. 1, is defective or short-circuited.

Should the entire test procedure prove that the vertical amplifier is all right then we know that the pentode

plate-coupling capacitator is opencircuited. To detect this condition a small signal voltage should be impressed across the input terminals of the vertical amplifier system. Then a vacuum-tube voltmeter is used to read the deflection voltage between the free vertical deflection plate and the chassis of the unit. Obviously, absence of any voltage between these points in the circuit shows that the coupling (C₀) capacitor between the vertical amplifier pentode plate and the free vertical deflection plate is open.

cathode potential.

Temporary Repairs

When the vertical amplifier gain control potentiometer is found defective, a temporary repair to the resistance element may be effected by connecting a fixed resistor, 1/4-megohin, between the slider arm and the ungrounded terminal of the resistance element.

Other defects which may affect the operation of an oscillograph are indicated by the failure of the horizontal deflection system to effect horizontal deflection of the beam. Here, the failure may involve either the vertical amplifier or the linear time-base oscillator system, or both. In the former instance, the testing procedure involved approximates that applied in the instance of the vertical amplifier. The testing is facilitated, however, by the provision for the direct connection of the horizontal amplifier input system to an external sweep-voltage source.

Capacitor Troubles

If the failure of the horizontal deflection system is confined to a single position of the linear time-base frequency coarse-range switch, the trouble will be traced to a punctured timing capacitor. In making the replacement, care should be taken to see that the replacement is of the same voltage rating and capacity rating as the original capacitor. A

temporary repair can be effected through connection of a nearly similar capacitor into this switch contact position.

The failure of the thyratron relaxation tube in the linear time-base oscillation circuit is readily detected through operation of the coarse-sweep frequency selector switch. Here, operation of the switch from one capacitor terminal to a succeeding capacitor effects a momentary deflection of the beam because of the rise in the uncharged capacitor potential from the zero to the maximum voltage level.

Timing Capacitor Defects

In several commercial cathode-ray oscillographs, the high frequency timing capacitor in the linear time-base oscillation circuit is permanently connected between the slider arm of the selector switch and the chassis of the device. Obviously, if this capacitor short-circuits, the entire system is incapacitated. This is indicated by the lack of a potential between the switch contact arm and the unit chassis. This check can be made with a d-c voltmeter having an internal resistance of 1000 ohms-per-volt, with the high-voltage rectifier tube removed from its socket

Synchronization Control Checks

An open circuit in the resistance element of the synchronization control potentiometer tends to render the linear time-base oscillator inoperative. Thus, if the shaft of this potentiometer is so positioned that the slider arm rests on the ungrounded portion of the resistance element, the control grid circuit of the relaxation thyratron is effectively open and the tube refuses to conduct and thus relax the charge on the timing capacitor. This simple test may be accomplished without removal of the oscillograph from its cabinet. A temporary repair may be effected by the same method used for the vertical amplifier gain control.

Wave Pattern Drifts

It is often observed that the image or wave pattern tends to drift along either the vertical or horizontal axis. This condition is an obvious indication that either the vertical or horizontal deflection plate series resistor (R₁₅ and R₁₆) in Fig. 1 is open circuited, or that either of the position control potentiometers is open. The latter condition is usually accompanied by more severe indications of trouble. Where the series deflection resistor is defect-

(Continued on page 44)

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Figs. 1 (left) and 2 (right) Fig. 1. A typical interstation noise suppression circuit. When no signal is present, the first audio tube is biased to cutoff. Upon tuning in a signal, the i-n-s tube is biased to cutoff, removing the excessive bias on the audio tube. Fig. 2. A typical i-n-s stage featuring two methods of controlling its effectiveness. The use of controls in the i-n-s circuit permits setting it to the appropriate point for local noise conditions



NOISE

INTERSTATION

A N inherent characteristic of f-m receivers is high interstation noise. That is, when tuning from station to station, where no signal is present, static, and other forms of interference are more pronounced than in equivalent a-m receivers. This condition is due to the wide-band acceptance of the f-m receiver. For this reason, most f-m receivers feature some form of interstation noise suppression or squelch system, popularly known as i-n-s.

How I-N-S Operates

The operation of the i-n-s stage is based on the biasing of an audio tube to cutoff when no station signal is being received. The circuit may be applied in one of three forms.

(1)—A switch may be used to cut the i-n-s system in or out of the receiving system.

(2)—Effectiveness of the i-n-s system may be adjustable.

(3)—The i-n-s stage may be permanently installed with no method of control.

Generally, a switch is incorporated in the i-n-s stage, so that its operation may be cut out for the reception of weak signals. This is necessary since i-n-s operation distorts the audio recovery of weak signals, or prevents their pickup entirely.

Typical I-N-S Stage

Fig. 1 shows a typical interstation noise suppression stage. The i-n-s tube is generally a high cathode-current type. The cathodes of the i-n-s tube and the first audio tube are connected to a common resistor. The control grid of the i-n-s tube is connected to the ave system of the receiver, or some source of rectified signal voltage. Since the cathodes of the two tubes are connected in parallel, when no ave voltage is applied to the control grid of the

by J. GEORGE STEWART

i-n-s tube, its cathode current is high, and the resultant bias voltage on the first a-f tube is of cutoff value. When a signal is received, the grid of the i-n-s tube is biased negative, reducing that portion of the cathode current supplied by the i-n-s tube, and restoring the effective bias on the audio tube to a normal value.

Switching

From the foregoing, it can be seen that unless the incoming signal is sufficiently strong to develop a high avc, or bias voltage on the i-n-s control grid, the a-f tube may operate at some high value of bias that may cause audio distortion. For this reason, most i-n-s circuits contain some form of switching system, which permits cutout and weak-signal reception. Other systems consist of both a switch and some form of control. Usually the Bsupply to the i-n-s plate or screen grid is varied or the effective bias on the a-f tube is varied, so that the system may be set to the appropriate operating point for local conditions. This method is illustrated in Fig. 2.

In this diagram, both methods of i-n-s action control are shown, although only one method is used in practice. Either method serves to regulate that portion of the bias voltage on the first a-f tube supplied by the i-n-s stage. The cathode type of control is preferred, since the characteristic action of the squelch is not affected as much.

Meissner 9-1053-4 A-M /F-M

In Fig. 3 appears the i-n-s circuit used in Meissner 9-1053-4 f-m/a-m receiver. A 6AC7 is used as the i-n-s

stage to squelch a 6SO7 first audio stage. With no station signal present. the bias on the 6SQ7 is 3 volts. This is sufficient to reduce the plate current to cutoff, and corresponds to a cathode current of 1.4 ma. With 90 volts on the plate of the 6SQ7, and a bias of 1 volt, its plate current is approximately .2 ma. For the 6SQ7 cathode resistor, 2,200 ohms, this would account for .44 bias volt, provided the cathode of the 6AC7 were not in parallel with it. Therefore, any avc voltage equal to or in excess of that required to reduce the cathode current of the 6AC7 to less than .3 ma., would open the first audio stage.

Potentiometer Uses

The switch in the cathode of the 6AC7 is used to cut out the i-n-s for reception of weak signals. To reduce the effectiveness of the i-n-s in localities where weak signals are received, a potentiometer across the screen-grid bleeder resistor, shown in dotted form in Fig. 3, could be adjusted so that the bias voltage on the 6SQ7 is reduced to some value below 3 volts. This voltage value should be selected so that no distortion is introduced when receiving weak signals, yet retaining some of the noise suppression.

Freed 57A-71A

Fig. 4 shows the i-n-s system used in the Freed model 57A-71A. Here, no switch is provided to cut out the i-n-s. Only one limiter stage is used in this receiver, and no avc voltage is applied to the r-f section. The rectified grid voltage of the limiter stage is used to actuate the magic eye, and operate the i-n-s. In the Freed model 40 F-M, a switch is incorporated in the Fig. 3. The i-n-s system used in the Meissner 9-1053-4. With no signal present, the bias on the 6SQ7 first audio tube is 3 volts, or cutoff bias. A fairly strong signal reduces the cathode current of the 6AC7, which in turn, reduces the bias on the first audio tube to 1 volt, or operating bias.



SUPPRESSION SYSTEMS

cathode of the i-n-s stage. For the average listener, weak-signal reception is unimportant, since any signal too weak to operate the i-n-s, would be too weak to operate the limiter. The resultant noise level, riding in with the signal would therefore make the signal unsatisfactory. Many receiver manufacturers therefore make the i-n-s system correlate its action with the limiter, so that the former operates properly for any signal which actuates the latter to its proper performance.

Fig. 4. The Freed 57A-71A á-n-s system. No cutout switch is used in this circuit, since any signal too weak to operate the i-n-s would be too weak for satisfactory listening.



Espey A-C / D-C F-M

In Fig. 5, a 14A4 is used for the i-n-s stage in Espey's f-m a-c/d-c receiver. Although the path of the actuating voltage is devious, it traces back to the limiter grid voltage developed across the bracketed resistors. Essentially, the circuit is identical to those of Figs. 3 and 4. However, the squelch action in this model is *soft*. That is, the noise is merely reduced to a comfortable level, so that the objectionable high level of interstation noise is materially reduced.

Checking I-N-S Stages

In checking i-n-s stages, it should be noted that their operation is dependent on d-c voltage characteristics. Where distortion is traced to this portion of the receiver, the bias on the first a-f tube should be checked, in the presence of an r-f signal. If this voltage is found to be high, the next step is to check the effective avc voltage being applied to the i-n-s control grid. The trouble will be found to be the resultant of improper control-grid or cathode voltages.

Adding an I-N-S Stage

Those owning f-m receivers without i-n-s stages may object to the high-interstation noise. This can be remedied by adding an i-n-s stage which includes but one tube, several resistors and capacitors.

In many f-m receivers, the first

audio stage is a single triode. For such receivers, a triode-pentode may be substituted, using the pentode portion for the squelch stage. This method is particularly adaptable to a-c/d-c type receivers, where the series filament voltage is unaffected by the introduction of a dual-purpose tube. It is suggested that when adding a squelch stage, its action be fixed. In this way, any objections to image responses in the receiver, by the set owner, are satisfied.

Fig. 5. An a-c/d-c type of i-n-s circuit, used in Espey f-m receivers. Because of the low available plate voltage, the action of this circuit is soft, that is, the interstation noise is reduced to a comfortable level.



SER-CUITS

to prevent unwanted absorption at certain resonant points. The r-f stage is impedance coupled to the 6SA7 detector; the plate choke must be a good one because of the three bands to be handled. The coupling lead is tapped down on the detector coils to reduce the load of the r-f amplifier, thereby increasing the Q of the circuit and, hence, its selectivity.

The oscillator uses cathode ticklers for the two lower-frequency bands and a tapped coil for the highest frequency. Selectivity is controlled in the first i-f transformer by connecting a tertiary winding in the secondary circuit for broad tuning. The second i-f transformer is similar but the tertiary is connected in series with the screen for degeneration. The audio amplifier is unconventional in the use of a 6SK7 first audio with the volume control following this stage. The audio tap from the rectified output of the detector is taken well down on the load resistor. across 70,000 ohms of a total of 470,-000, which sacrifices signal voltage but improves the detector linearity. A decoupling filter of 30,000 ohms and .15-mfd acts to prevent feedback.

A 2-megohm tapped volume control feeds a 6P5G driver. A 6SQ7 picks off some of the driver's output for phase inversion to feed the push-pull 6V6 output. One of the diodes serves as the second detector, the other is grounded. The tone control is conven-

(Continued from page 34)

tional. The driver cathode is connected to the voice coil of the speaker in radio operation, but not on phono, for introducing degeneration. The voltage is taken from the midpoint of an 8000ohm potentiometer. The d-c bias value is effectively two-4000 ohm resistors in parallel, or 2000 ohms with no bypass capacitor.

Firestone S-7406/3

The Firestone S-7406/3 is another example of interesting high-fidelity design. This is a 3-band receiver with an electrostatically shielded loop, tuned r-f stage with a 3-gang capacitor and four 6SQ7s for detector, avc and audio. This model uses a separate 6J5 oscillator with grid-to-grid coupling to a 6SA7 mixer. A standard 6SK7 i-f amplifier feeds a single diode of the first 6SQ7 which acts as detector only. Avc is obtained by rectifying the voltage picked directly from a 6SK7 plate through a small capacitor by the diodes of a second 6SQ7. A third 6SQ7 serves as first audio while a fourth acts as a phase inverter. The tone control is tied to

Fig. 2. Portable Motorola battery a-c/d-c receiver. An untuned r-f coupling transformer is used to increase the gain of the receiver. The coupling in this transformer is adjustable through the use of a 5-mmfd gimmic. the volume-control bass compensating circuit, so that with the grounded arm at the top, a capacitor is placed across the volume control to reduce highs while with the arm at the bottom, the shunting effect of the capacitor is removed and the bass compensating capacitor is cut out. This reduces the bass.

Motorola 65BP 1-2-3-4

In Fig. 2 we have a battery a-c/d-c high fidelity portable, Motorola 65BP 1-2-3-4. A 5-mmfd gimmic is used to increase the coupling in the r-f coupling transformer, thereby increasing the gain of the receiver. This gimmic consists of a stiff piece of wire connected to the secondary of the r-f transformer, whose coupling to a small coil, connected to the primary of the same transformer, may be varied. There is no physical connection between the two parts. The effective capacitance is estimated at 5 mmfd.

A novel method is used to bias the 3Q5 power amplifier. The control grid of the 3Q5 is returned to terminal 7 of the 1H5GT. This places the filament of the 3Q5 at 4.5 volts positive with relation to the control grid. The 720-ohm resistor and 20-mfd capacitor is used to provide additional filtering for this bias voltage. The additional current drain through the 3Q5 filament is negligible.



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OSCILLOGRAPH SERVICING

(Continued from page 38)

ive, it is customary to replace both series resistors.

Where the vertical or the horizontal potentiometer fails to center the pattern written on the screen, despite complete rotation of the control shaft, we usually find a leaky or short-circuited deflection plate-coupling capacitor in the coupling circuit. Again, if the tube fails to write the expected pattern and a fluorescent glow is observed along an edge of the screen when the intensity and focus potentiometers are advanced in a clockwise direction, the trouble will be found to be an open resistance element in the vertical or horizontal-positioning control potentiometer.

Lower Edge Fluorescence

A green or fluorescent glow along the lower edge of the fluorescent screen indicates that the positive side of the vertical-control potentiometer resistance element is open. Counterclockwise rotation of the potentiometer shaft causes the image to move rapidly in an upward direction, leaving a fluorescent glow at the upper edge of the electron screen. Thus, a fluorescent glow at the extreme left-hand position of the screen indicates that the horizontal-control potentiometer resistance element is open circuited, and that the slider arm of this potentiometer rests on the negative portion of the resistance element. The position-control potentiometers incorporate high resistance elements, and should therefore be replaced immediately.

Inability of the focussing potentiometer to properly focus the beam on the fluorescent screen is obviously an indication that the potentiometer is defective. Operation of the potentiometer results in the development of a diffused glow over the entire surface of the fluorescent screen once the slider arm of the potentiometer rests on the positive portion of the resistance element. The potentiometer, in this instance, should be replaced immediately.

Intensity Control Opens

An open circuit in the resistance element of the intensity-control potentiometer precludes the possibility of the development of a screen pattern, as long as the slider arm of the potentiometer rests on the negative portion of the resistance element. Operation of



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Be Right with OHMITE RHEOSTATS • RESISTORS • TAP SWITCHES the potentiometer in such a manner that the slider arm contacts the positive portion of the open-circuited resistance element results in the development of a bright, suffused screen glow, with attendant inability of the focussing control potentiometer to properly focus the electron beam on the fluorescent screen. The transition from the former to the latter condition occurs abruptly with the counter-clockwise rotation of the intensity control potentiometer control knob. When this condition occurs, the defective potentiometer should be replaced, inasmuch as a temporary repair with a fixed resistor will not permit the fine, continuous control of the beam intensity.

General Servicing Problems

In general, servicing of the cathode-ray oscillograph is similar to receiver servicing. The general precaution of removing the high-voltage rectifier tubes when making vertical or horizontal deflection system voltage tests must be observed. Hence, necessary testing of the high-voltage circuit components should be carried out with the proper ohmmeter or component tester, whether the latter be an inductance or a capacitance tester. All d-c voltage measurements should be carried out with units having an internal resistance of 1000 ohms-per-volt, since the resistances incorporated in the amplifier and deflection circuits are very high. Finally, if there is definite indication, from preliminary pattern analysis that one of the c-r potentiometers is open-circuited, the highvoltage filtering capacitor must be discharged. This is accomplished by short-circuiting the plate and one of the heater contacts directly at the high voltage rectifier tube socket.

When the improper operation of the oscillograph necessitates the replacement of one of its components, the replacement should be effected with a component of equal or better quality.

RAILROAD RADIO ANTENNA SYSTEM



Half-wave antennas used in an 118-mc f-m system aboard the Denver and Rio Grande Western R.R. (Courtesy Galvin Mfg. Co.)



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AUTOMATIC NOISE LIMITERS

(Continued from page 15)

RC network of the infinite impedance detector is very fast, as will be noted from the parts values.

RME 41-43

In Fig. 5 we have the a-n-l circuit of the RME 41-43. This system is an adaptation of the automatic a-n-l system shown in Fig. 1c. Here, again, the percentage modulation is used to control noise pulses. Since the cathode of the noise limiter is at a higher positive potential than the plate of the same tube, no conduction takes place. However, the RC network in the feeder line to the plate of the limiter diode keeps the voltage on the plate at carrier level, while the cathode voltage of the same tube varies with the modulation. When the modulation exceeds 100%, the voltage on the plate of the noise limiter becomes positive with relation to its cathode, and the tube acts as a low-impedance path across the audio feeder network. We note that the 1-mfd capacitor returns the a-n-l plate to the ground side of the audio network, thereby placing the diode across the audio load. The insert in Fig. 5 shows the circuit reduced to a bridge network to further explain its action.

Servicing Problems

Component value plays an important role in a-n-l action. The usual troubles encountered in their servicing involves either no action or distortion of the signal. Component values should be checked closely, and replacement should always be made with identical values. Since the cathodes of the a-n-l tubes are at r-f or a-f potential, they are a frequent source of hum caused by cathode to filament shorts. The tubes used in the a-n-l stage should always be checked where hum conditions arise. Tt should also be noted that a-n-l systems are most effective where the signal strength is high. Weak signals will usually prevent adequate noise limiting.



F-M ANTENNA

The f-m antenna of the Miami police department.



SUB-MINIATURES

(Continued from page 20)

are comparatively much larger in size. The new tubes measure approximately $1\frac{1}{2}$ " in height (neglecting prong length), are oblong in diameter with the longer side about 7/16" and the shorter less than $\frac{3}{8}$ ". They are available in two styles, based or with flexible leads.

Hearing-Aid and Sub-Miniature Similarities

These new tubes are almost identical in appearance to those used in hearing aids. In both styles of mount, the leads come out of the glass envelope in a straight line, with a polarizing mark on the glass envelope. Sockets for use with these new type tubes are also polarized. At present, four types are available; triode-heptode for converter use, sharp cut-off pentode for use as a r-f amplifier, diode pentode for detector first audio applications, and a power pentode for output circuits. Thus, a complete superheterodyne re-

(Continued on page 48)



Figs. 6 (above) and 7 (below). In Fig. 6 appears a Belmont pocket receiver using the new sub-miniature type tubes. Receiver is 3" wide, 34" thick, and 614" high. Weighs about 10 ounces, including batteries. Fig. 7 shows a cross-sectional view of the sub-miniature tube used in the receiver.





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6 TUBE AC-DC KIT



SUB-MINIATURE TYPE TUBES

(Continued from page 47)



ceiver may be designed around these tubes.

As shown in Fig. 2, the filament voltage for all types is 1.25 volts at current drains varying from 30 to 50 ma. The use of 1.25-volt filaments permits their use from $1\frac{1}{2}$ -volt sources to an end voltage of 1. Raytheon types operate at plate voltages ranging from $22\frac{1}{2}$ to 45, while the Sylvania types can be used with plate supplies up to $67\frac{1}{2}$ volts.

Fig. 3 shows the comparative characteristics of sub-miniature, miniature and standard size filament type r-f pentodes. It will be noted that the gain for the sub-miniature type compares favorably with the others, considering the low plate and filament voltages and currents. Circuit applications are identical to those for miniature types, as we note in Figs. 1 and 8. In Fig. 1

Fig. 9. Components used in receiver diagramed

above.

Fig. 8. Circuit of a sub-miniature type receiver using a 1T4 r-f, 3401 detector, 3402 a-f and 3403 output. The 3400 type tubes were made by Microtube Laboratories.

we have a typical amplifier circuit using a double space-charge tetrode voltage amplifier. Fig. 8 illustrates a complete receiver using sub-miniature tubes and components.



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AVC/DETECTOR/A-F SYSTEMS

(Continued from page 24)

.005-mfd capacitor and a $\frac{1}{2}$ -megohm rheostat, is conected across the triode output.

G. E. 20

An example of how an extra amplifier stage is used to make up for the loss in gain due to degeneration is shown in Fig. 5; G. E. 20. The 6SQ7 amplifier, similar to the one in Fig. 4, feeds a 6J5 second or intermediate audio stage which, in turn, feeds a 25L6 output tube. A tertiary winding on the output transformer supplies the feedback voltage which is fed to the cathode of the 6J5 through a 3300ohm bias resistor. The grid leak of the intermediate stage serves as the variable tone-control resistor. A phono jack is connected directly to the first audio grid.

Ward 62-271

A circuit using a 6C5 triode as a diode with the grid as plate, the plate being grounded, is shown in Fig. 6; Ward 62-271. This detector has an unusually high load resistor. .3 megohm, and, hence, a low value capacity shunt, .50 mmfd. A 2-section i-f filter with two .1-megohm resistors precedes a 2-megohm tapped volume control. This arrangement constitutes an a-c load potentiometer with the 6C5 first audio grid tapped down .2 of a megohm which reduces the a-f loading and, consequently, the distortion. The volume control, which is directly in the first a-f grid circuit, has a 25mmfd capacitor shunting the high end, providing a boost to the treble frequencies. This may be considered an equalizer for flattening the audio response. An .08-mfd capacitor and 7000-ohm resistor shunt the low end of the control for bass compensation. An enforced cathode bias is applied to the first audio instead of the usual cathode resistor. This is accomplished by bleeding some d-c from the B supply through a potentiometer consisting of 25,000 ohms, from B+ to cathode and 400 ohms from cathode to B---(ground). An audio bypass of 12 mfd shunts the 400-ohm unit.

RCA QB1

A 6-volt 150-mil tube farm receiver circuit appears in Fig. 7; RCA model QB1. A series resistor ahead of the

(Continued on page 50)



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AVC/DETECTOR/A-F SYSTEMS

(Continued from page 49)

volume control for r-f filtering and reducing a-f loading, is used here, too. The second filter capacitor consists of a .01-mfd audio coupling capacitor in series with a 100-mmfd capacitor shunting the 6T7 grid. Note the bias cell and 2.2-megohm grid leak. A .0025-mfd tone capacitor shunts the first audio output.

RWT B92

An extreme example of reduced detector loading is used in the circuit of Fig. 8; Radio Wire Television B92. Considerable detector output may be sacrified because a 6B7 double-diode pentode is used, followed by a 41 connected as a triode driver for a 6A6 class B output amplifier. A degenerative link begins at the voice coil and runs through a .01-mfd capacitor to the first audio plate. The diode load consists of a .25-megohm resistor and a .2-megohm volume control. Both high boost and bass compensation are featured. A separate resistance/capacity filter of 20,000 ohms and a .25 mfd supplies the 6B7.

National NC-44B

In communications receivers it is customary to provide an avc-cutoff switch for manual volume control, particularly when using a beat oscillator for c-w reception. The beat oscillator is usually connected to the second detector, as in the National NC-44B; Fig. 9. With avc working, we have a continuous r-f detector input with the consequent high avc bias, killing the sensitivity of the receiver. When avc is turned off, one or more r-f stages may be overloaded by strong signals. Thus an r-f gain control is provided in addition to the usual a-f control. In this receiver, cathode bias is used on the 6SQ7 instead of the usual grid-leak bias.

Motorola 41H

A circuit using a 3A8GT combination tube for i-f, detector and a-f appears in Fig. 10; Motorola 41H. A single diode detector is fed by the pentode section serving as i-f amplifier. No ave is fed to this section which has a 2.2-megohm grid leak anti-blast bias. This system is usually restricted to receivers having little pickup. The 1/2-megohm volume control acts as the diode load resistor with a .01-mfd coupling capacitor and a 3.3-megohm leak for the triode first audio section.

VOLUME AND TONECONTROLS

(Continued from page 30)

others. When selecting the proper control for audio-grid circuit replacement work, it is wise to try to choose a unit having not too abrupt a change of taper at this region.

Caution: It is entirely possible that in many cases the replacement control specified by the replacement control manufacturer's list, manual, or guide will not correspond exactly in type, taper or even in resistance value with the faulty unit found in the receiver. The reason could well be that the receiver was serviced for volume or tone-control trouble previously by another Service Man and, due to the war-time shortage of controls (or some other procurement difficulty) that existed at the time, a control of incorrect specifications was installed as an expedient to get the set working, even though an entirely satisfactory volume or tone control was not provided in this procedure.

When such cases are encountered, it will be advisable to use the replacement unit recommended in the replacementcontrol manufacturer's manual or guide, rather than to simply insert a control identical with the faulty one. A study of the type of control circuit employed in the receiver will serve as a check in case of doubt, since the new unit to be installed should be of the type and specifications to fit the requirements of the particular type of control circuit and tubes employed.

High-Voltage Couplers for Controls

Standard controls are usually supplied with the center terminal and moving member insulated from the mounting bushing and case. For applications where controls must be used in high-voltage circuits (such as in television equipment, cathode-ray oscillographs and other electronic applications), one manufacturer has provided a high-voltage insulating coupler in the control assembly. As shown in the two illustrations of Fig. 5, this feature takes the form of a bakelite spacer tubing that screws on to the regular control mounting bushing at one end, and carries a metal bushing through which a shaft protrudes at the other, with an enclosed bakelite shaft coupling joining this shaft and the slotted short shaft of the control. The spacercoupler can be of any required length to withstand the existing operating voltage, but the two types available at present will withstand an insulation breakdown of either 3,000 or 10,000 volts a-c between the control proper and its shaft and mounting.

This type of coupler is now being made available with certain types of midget composition-element controls



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and with wire-wound and multipletype controls.

SOCKET PREHEATING

In the next article of this series, repair and reconditioning of volume and tone controls will be explained.

(To be continued)

At right, preheating tube device used at General Electric to check tubes in warmed-up condition and thus determine in-use efficiency. Correlated data also serve to assist the development engineer in improving tube design and construction.





SERVICING HELPS

Sparton 410, 420

No reception from 550 to 800 kilocycles, reception above 800 kc erratic at times: Condition due to dirty wiper contacts that ground the rotor of tuning capacitor. Shaft and wiper should be cleaned with carbon tetrachloride, applied with a small art brush, rotating the condenser while applying.

Philco 112-112A

Inoperative: Usually the input or output transformers have open windings. Of course the proper way is to replace the transformer. But, if the customer is willing, a resistor can be placed across the open winding, and a capacitor from that point to the other winding will effect a very satisfactory repair.

Synchronous Motor Record Players

Wavering of speed, especially on slow music: While cleaning and oiling will improve operation, the condition may reappear. Thus it has been found better to use the following method, involving a small repair to the center shaft. This shaft, which is fixed to the turntable, works on a single ball bearing. The ball bearing wears a spot in the shaft. Enough of the shaft should be ground off to remove all trace of the spot. Shaft should then be cleaned, oiled and replaced.

New Dial Drive Installation

New cables tend to slip on shaft, even though there is sufficient spring tension: To cure, a small amount of resin should be powdered on the cable. To provide a base for the resin carbon tetrachloride or alcohol can be used. A very small amount should be applied or the cable will follow shaft. If too much has been applied, brushing with carbon tetrachloride or alcohol over the shaft and cable will help.

Claude M. Prew

G. E. 60

High-pitched audio squeal: Traced to volume control. The control should be cleaned with carbon tetrachloride. A piece of No. 30 wire should be inserted under the U clamp on the shaft to take up any play. The carbon strip should be lubricated with graphite (lead pencil).

350-SPEAKER P-A System

A 350-SPEAKER p-a system was recently installed in the Rochester plant of Stromberg-Carlson. Complete control of the system is achieved with a master console. This unit controls a two-channel system for either paging or music to thirteen groups of speakers throughout offices and factories. A third channel, for recording, may be used when each of the other two channels is in use.

In addition to the thirteen groups of speakers the new console controls thirty power amplifiers, eight microphone locations using ten microphones, a radio tuner, and a transcription player.

It is possible to control the broadcasting of music throughout the plant while permitting the operator to page any desired section without affecting the music in the others.

The input panel controls, by key switches, linking of ten microphones on either paging, music, or recording pre-amplifiers. A key switch is provided to connect the transcription player to either the radio music or the recording pre-amplifiers. Keys are also provided for fire, emergency, or time signals. Radio is connected by a key switch. Seven individual mixers enable the operator to control the level. Eight telephone circuits connect the console with the local plant telephone switchboard, so that paging calls can be handled effectively by the sound system operator.







At left, a view of the console and the glass-partitioned studio used for special employee plant broadcasts. To the left of the console are the recording pre-amplifiers and additional key switches. Above appears an overall view of the console, which includes the broadcast receiver. Programs picked up by the receiver are also piped into the console and over the p-a plant system.



by SERVICER

OLD TIMER'S

A BOUT the most debated topic in town today is surplus. Seems as if everyone has something to say about it. We had quite a session on the subject a few nights ago during one of our weekly meetings. Roy started the ball rolling by telling us how keen he was about surplus parts. He had just returned from the Big City, where he had visited a surplus warehouse.

"I tell you," gushed Roy, "it was the sweetest and the most complete assortment of parts I have seen in a long time."

"Did you buy some?" asked John.

"Yep, I did," said Roy. "And what's more I think that I shall be in a position to do some fancy buying from now on. Wait till you see the hard-to-get parts I pick up!"

"How do you figure that?" asked John. "You don't think that you can really use those parts in just the same condition in which you bought them, do you?"

"Well," continued Roy, "there is a thing to be done here and there to make them work. But for the most part they will fit with a bit of tailoring. And I for one would rather repair than not repair, if you get what I mean, and I think that you do!"

Pete's Argument

"You boys don't know what you are talking about," broke in Pete. "I've been listening to you tell how you are going to make a killing and have parts that are hard to get."

"Don't you agree?" asked Eddie. "At least Roy here will surely have parts!"

"Sure he'll have parts," continued Pete, "but what kind of parts?

"This idea that surplus is the answer to your problems is all wrong."

"And how do you figure that all out," asked Roy.

"Well, it's true there's quite a surplus load about. We find everything from ½watt resistors, and maybe even smaller ones, to complete Navy and Army receivers. And to be sume it should be sold to someone. But we shuldn't buy !

"The manufacturers are now just getting under way. They will be bringing out parts that will fit and parts that will go into the greatest majority of sets now on the market and to come on the market after a little while. It is expected that the Service Man will want to get his repairing down to a science and that he will want to repair each same set the same way with the same type of parts.

Tolerance Problems

Tolerances of the surplus pieces are going to vary all over the map. Most sets can be fixed with 20% resistors and 20% capacitors. That is what we have been using since radios were brought into us for repair. The surplus stuff *might be* 20% or rejects with broader plus or minus ratings.

"You are not going to be able to tell which was manufactured yesterday or just before the end of the war, and what was manufactured sometime late in 1941 or the beginning of 1942. So you won't know the part's shelf life. And don't tell me that that is not important. We all know that it makes a great deal of difference if we use a fresh electrolytic or one that has been around for a few years.

Guarantees

"Then there is the matter of guarantees. Which surplus material is guaranteed under the standard RMA guarantee? None! And what are you fellows going to do when something pops and you want the part replaced by a manufacturer? Sure, I know, that doesn't happen very many times, but when it does, it certainly is a nice feeling to know that not everything is lost in the transaction.

Manufacturer's Helps

"And then there's another matter which I throw before you for consideration. Are we as Service Men really helping the manufacturers as much as we would want them to help us? How can we ask them to take the trouble to help us with our technical problems, which are many, and which we blithely throw into the respective laps of the tube manufacturer, the resistor manufacturer and the capacitor manufacturer? Is it fair to buy surplus and then if the part's don't work, ask a manufacturer why they don't? Or should we show our loyalty by staying with the firms that helped get us established. I think that while we are debating that the manufacturer's should be backing us up, we should be supporting the manufacturers with our business.

Credits

"And last but not least," concluded Pete, "we can better establish ourselves on the credit side of the ledger by dealing with the first-line manufacturers through our distributor. Once we have that credit established, then we can sail along without much trouble through good times and bad. All of that we lose by splurging in surplus.

No Deal for Surplus

"That's why I think that surplus offers no deal for us. I, for one, am buying parts made by old-time manufacturers. They will come through in time for me, I'm sure. And they won't let the rest of you fellows down either."



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INTER-CITY TELEVISION SYSTEM



How recent television programs have been broadcast from city to city. At left, above, appears the Army-Navy football game setup from Philadelphia. At right, above, appears the link used to transmit from Washington, D. C., to New York and over three stations in the New York area. Similar links will be used throughout the country as television transmitters are installed.



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RIEDEL RESIGNS FROM RAYTHEON

Edgar S. Reidel has resigned from the Raytheon Manufacturing Company. For 13 years he was general sales manager of the receiving tube division. Mr. Riedel was one of the nine original organizers of the RMA at Chicago in 1923.



JESTER, KARET AND OGILVIE WIN MAGUIRE PROMOTIONS

Oden F. Jester has been named general sales manager of the radio and phonograph division of Maguire Industries, Inc.

Mr. Jester will direct sales of Meissner radiophonographs and other products to be made by the Meissner manufacturing division at Mt. Carmel, Illinois, record changers and similar products of the Maguire plant in Bridgeport, Conn., products of the Thordarson Electric manufacturing division in Chicago and of the Radiart Corp. subsidiary in Cleveland, Ohio.

Robert M. Karet has been appointed manager of the electronic distributor and industrial sales department of Maguire Industries.

Under Mr. Karet's direction, all distributor and industrial sales of Maguire subsidiaries serving the electronic field will be coordinated.

Mr. Karet will make his headquarters at 936 North Michigan Avenue, Chicago.

Allan R. Ogilvie has become a vicepresident of Maguire Industries and placed in charge of the Bridgeport, Conn., plant. Mr. Ogilvie was formerly chief en-

ton Wasmansdorff succeeds Mr. Ogilvie.

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12

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AIREON PIEZOELECTRIC CRYSTAL CATALOG

A crystal catalog has been published by the Aireon Manufacturing Corporation, Kansas City, Kansas.

by the rine constantiation in the corporation, Kansas City, Kansas. Principal types of crystals described in the catalog are: Octal type with cylindrical metal shield and standard eight pin base; three-pin, two-channel, aircraft type; standard two-pin phenolic holders for various kinds of mobile and stationary installations (banana or pin plugs); and variable air-gap mounting with screw top electrode.

OPERADIO IMPEDANCE CALCULATOR

An *impedance calculator*, 5" in diameter, designed to aid in matching loudspeaker lines to an amplifier for sound systems with 500, 1,000, 4,000, 8,000 or 16,000-ohm loudspeakers, has been developed by engineers of Operadio Manufacturing Co., St. Charles, Illinois.

Scales for 500, 1000, 4000 and 8000ohm loudspeakers are incorporated on three revolving discs. Total group impedance of all loudspeakers is obtained by matching specific ohm scales to the number of loudspeakers involved in a sound installation. Available to sound men ior 25¢, direct or from distributors.



L. J. WRONKE JOINS HALLICRAFTERS

Louis J. Wronke, Inc., Oak Park, Ill., industrial designers and engineers, has merged with the Hallicrafters Company.

The Wronke Corporation's president, Louis J. Wronke, has joined Hallicrafters as chief mechanical engineer and director of design.

ALLIED RADIO LAUNCHES RAYTHEON BONDED PLAN

Allied Radio, 833 W. Jackson Boulevard, Chicago, Ill., announced adoption of the Raytheon bonded plan at a recent dinner meeting of dealers. A. D. Davis, president, and W. F. Marsh, sales manager, analyzed the plan. (Full details (Continued on page 58)



• Now, more than ever before, you just can't afford to take a chance with questionable capacitors. In other words, be sure that the capacitors you install are "going to stay put."

And that's precisely where Aerovox capacitors come in. They're fresh. The popularity of the Aerovox line means quick turnover of jobber stocks. You can be sure that Aerovox capacitors are of current production. And that's especially important in electrolytics. Don't take a chance with unlabeled or questionable capacitors!

• Ask Our Jobber . . .

Ask about the various types of capacitors you need. Ask for copy of our latest catalog—or write us direct. Meanwhile—look for that YELLOW label, which means Aerovox, and Aerovox means DEPEND-ABLE capacitors.



In Canada: AEROVOX CANADA LTD., HAMILTON, DNT. Export: 13 E: 40,ST., New York 16, N.Y.- Cable: 'ARLAB'



NEWS

(Continued from page 57)

of the plan appeared in the November issue of SERVICE).

GENERAL CEMENT DIAL BELT DISPLAY

A dial-belt display stand has been announced by the General Cement Manufacturing Co., 919 Taylor Avenue, Rockford, Ill.

NEWS OF THE REPRESENTATIVES

O. P. Smith, 660 N. Dearborn St., Earl T. Champion, 4753 N. Broadway, and Gordon E. Gray, 1 N. Pulaski Rd., have become members of the Chicagoland chapter of the Representatives.

A. B. Smedley, 15 N. Euclid Ave., Pasadena, Cal., has become a member of the Los Angeles chapter.

At the annual election of the Pacific Northwest chapter, Lloyd Marsh was elected president; Ralph James, vice president; and George Norris, secretarytreasurer.

John J. Hagerty, 1010 Francis Palms Bldg., Detroit, Mich. was recently accepted as an associate member of the Wolverine chapter.

The following members of the Chicagoland chapter have moved: J. E. Goode, 412 So. Green St., Chicago 7; Bob Whan, 4717 Madison St., Chicago 44; and Wm. J. Johnston, 549 W. Randolph St., Chicago 6.

REMOTE CONTROL TELEVISION ANTENNA



Remote control antenna system, developed by Farnsworth for television reception, to eliminate ghosts or shadows that are apparent on receivers in large cities where skyscrapers are numerous and in mountainous regions such as in California. Within the case are two motors, one for orientation and one for tuning. Four push buttons on a control board allow for the receiver operator to rotate the antenna elockwise or couterclockwise through 180° and to increase or decrease the frequency of resonance. A deluxe installation has a three-section arm extension with a frequency range of 2½ : 1, from

A deluxe installation has a three-section arm extension with a frequency range of $2V_2$: 1, from 46 to 117 mc. Smaller antenna system has two sections with a frequency range of 1.8 : 1 or 46 me at bottom and when extended, to 85 me at the top.





TRIPLETT VOLT-OHM-MILLIAMMETER

A volt-ohm-milliammeter, type 2405, with 25,000 ohms-per-volt (d-c), has been announced by the Triplett Electrical Instrument Co., Bluffton, Ohio. Ranges are: D-c volts 0-10-50-250-500-

1000, at 25,000 ohms-per-volt; a-c volts ... 0-10-50-250-500-1,000, at 1,000 ohms-pervolt; d-c amperes ... 0-10; a-c amperes. 0-0.5-1-5-10; d-c milliamperes...0-1-10-50-250; d-c microamperes ... 0-50; ohmsmegohms ... 0-4000-40,000 ohms/4-40 megohm; output...capacitator in series with a-c volts. Has a 6" model 626 micro-







ammeter, adjusted to 40 microamperes. Metal case, $10'' \ge 534''$. Weight, approximately 11 pounds. Leather strap handle for portable use. Batteries selfcontained. *

STRUTHERS-DUNN REVERSING CONTACTOR

Heavy duty reversing contactors, type 135CXX, have been produced by Struth-ers Dunn, Inc., 1321 Arch Street, Philadelphia, 7, Pa.

Designed for use with polyphase motors up to 1 h-p, single-phase motors to 3/4 h-p. and is available for from 110 volts to 600 volts a-c. Consists of two 3-pole contactors mechanically interlocked. For control operation on door and window operating devices, machine tool cross feeds, etc. Size: 534'' wide x 4'' high x 178'' deep. Weight, approximately $2\frac{1}{2}$ pounds.



DUMONT 7" TELEVISION TUBE

A 7" television receiver tube, type 7EP4, has been developed by the Allen B. Du Mont Labs., Inc., Passaic, N. J. Provides for a normal screen image of 4" wide by 4!4" high. Tube is 15!/2" long. Accelerating poten-

tial is 2500 volts.



G.E. MIDGET THYRATRONS All-metal midget thyratrons, type GL-502A, have been announced by the tube division of G.E. The GL-502A is an inert-gas-filled, (Continued on page 60)





This compact, inexpensive, constant- impedance output attenuator (Clarostat Series CIB) will dissipate 10 watts at any setting. Operates noiselessly without distortion. It is highly recommended as an individual speaker control in a multi-speaker P-A system. Linear attenuation in 3 db steps up to 30 db, and then final step to infinity. Insertion loss is zero.

The Clarostat standard line (listed in latest catalog) also includes wire-wound T-pads and L-pads of the constant-impedance type for use in sound systems where the associated equipment in circuit must remain within the limits of a required constant value.

★ Ask Our Jobber...

Ask him about Clarostat sound-system controls, as well as other controls and resistors you require. Ask for latest catalog -or write us direct.



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> Yes. Vaco has created more than just a variety of screw drivers. Vaco has built the exact type of screw driver to dype particular job that can be tedious and troublesome when an ordinary driver is used. No wonder mechanics who do precision work say Vacos are "tops" among all drivers. Vacos, with gleaming Amberyt handles, are shock-proof and breakproof. Write for catalog.

ACO TYPES



Canadian Warehouse: 540 King St. W., Toronto, 2

NEW PRODUCTS

(Continued from page 59)

double-grid thyratron with negative control characteristics. The control characteristic is said to be independent of ambient temperature over a wide range. Net weight, two ounces. Height, $2\frac{1}{2}$, diameter, $1\frac{5}{6}$.

SIMPSON TUBE TESTER

A mutual conductance tube tester has been announced by the Simpson Electric Company, Chicago.

Features a method of testing tubes in freatures a method of testing tubes in terms of percentage of rated dynamic mutual conductance. Tube under test is compared with the standard rated micromho value for that tube. Colored zones on the dial coincide with the micromho rating or percent of mutual conductance, indicating that the tube is good, fair, doubtful or definitely bad.

Has ten push-button switches and nine rotating switches of six positions. An automatic re-set button returns all switches to normal when the test is completed. Tube chart is provided. Overall size, $15\frac{1}{2}$ " x $9\frac{1}{2}$ " x $6\frac{1}{2}$ ".



CLARK 30-WATT AMPLIFIERS

Thirty-watt amplifiers have been announced by the Clark Radio Equipment Corporation, 4313 Lincoln Avenue, Chicago, 18, Ill.

Features include: Three input circuits, one high impedance (500,000 ohms) phonograph, plus two high impedance (500,000 ohms) microphone channels; microphone gain is said to be 125 db, phonograph 85 db; two speaker sockets with impedance tap switch providing 4, 8, 16, or 500 ohms; individual bass and treble boost or attenuation (bass equalizer variable, said to be from -12 db at 50 cycles to +18 db at 70 cycles, treble equalizer from -12 db to +21 db at 10,000 cycles);



irequency response said to be ± 2 db from 50 to 10,000 cycles; hum level said to be about 55 db below maximum output with bass control on full.

Tubes: 3-7C7s, 1-7F7, 1-6SN7, 2-6L6s, 1-5U4. Power consumption: 150 watts at 117 volts, 60 cycles.

117 volts, 60 cycles. Dimensions: 10'' high, 20'' wide including handles, 115%'' deep.

SHALLCROSS 1.5-WATT HERMETICALLY-SEALED RESISTORS

Hermetically-scaled 1.5-watt resistors, type 1105, have been announced by the Shallcross Mfg. Company, Jackson and Pusev Avenues, Collingdale, Pa. Size $2 1/16'' \log x 7\%''$ diameter. Maximum ratings are 1 megohm resistance and 1,000 volts. Wound with nickel chromium wire.



G. E. INDUSTRIAL SOLDERING IRONS

Industrial soldering irons, ranging from 75 to 300 watts, with $\frac{3}{8}'' \ge 1\frac{1}{4}''$ tips have been announced by the industrial heating division of G.E.

Irons are said to have quick recovery and high reserve-heat capacity. Parts subjected to high temperatures use calorized (surface-alloyed with aluminum) copper and 18-8 stainless steel. Heating units are replaceable.



E. F. JOHNSON TO MAKE PLUGS AND DIAL LIGHT ASSEMBLIES

The E. F. Johnson Company, Waseca, Minnesota, has acquired all tools, inventory and manufacturing rights for Mallory-Yaxley cable connectors, pilot and dial light assemblies, tip plugs and tip jacks.

Seven- and twelve-wire cable poralized connectors will be produced. They will be



available with several types of mountings

for both the receptable and pin plugs. Dial lights will be supplied as shell assemblies and with slip-on brackets. Tip plugs will be of the solderless type and supplied in a long and short length. Tip jacks will be available with either metal or bakelite type heads and with round and hexagon heads. Twin tip jacks are molded bakelite and will be available in two types, shorting and non-shorting.

* * * OPERADIO RAILROAD LOUDSPEAKERS

Railroad-type loudspeakers engineered for railroad traffic control or intercom-munication have been developed by Operadio Manufacturing Co., St. Charles, Illinois.

Has a pressure neutralizing grill and filter, which is said to eliminate air pressure on diaphragm when loudspeaker is used on the exterior of an engine or caboose. Also has stream filtration to





New MODEL 450 TUBE TESTER

Designed especially for the service man who wants to give better service. Speedy operation is assured by newly designed rotary selector switch which replaces the usual snap, toggle or lever action switches. Tests all tubes up to 117 volts including 4, 5, 6, 7L, Octals, Loctals, Magic Eye, Hearing Aid, Thyratrons, Single Ended, Floating Filament, Mercury Vapor Recti-fiers, also Pilot Lights. Tests shorts and leakages up to 3 Megohms in all tubes. Tests both plates in rectifiers. Has many other new features and ad-vantages that ang good ser-vice man will appreciate af-ter five minutes use. Immer **\$3050** diate Delivery.

Write for latest bulletin showing hundreds of other hard-to-get items at money-saving prices.



avoid accumulation of soot or dirt. Speaker housing is sealed and weatherproofed.

WELLER SOLDERING GUN

A transformer type electric soldering gun, model B, which is said to come to temperature in 5 seconds, has been developed by the Weller Manufacturing Company, Easton, Pa.

Has a trigger switch. Current is broken the minute the finger is removed from the trigger and the tool is laid on the bench. Consumes 100 watts; operates on 115 volts, 60 cycles.



ANGLO CONSTANT-SPEED D-C MOTORS

Constant-speed d-c motors especially adaptable to synchronous operation have been announced by the Anglo Corporation, 4234 Lincoln Ave., Chicago.

Motor is self-starting. Current con-sumption is from .06 to 1 watt. Shaft speeds may be geared from 1 revolution every 24 hours up to 600 rpm.

Available for use at 11/2, 3, 6, 12, 24, 32 or 110 volts.



PARAGON TIME SWITCHES

Time switches, 300 series, with telechron motors, have been developed by the Para-gon Electric Company, 37 West Van Buren Street, Chicago 5, Illinois.





Made right . . . to work right . . . and stay right. Whether in stock ratings or to your own specifications you will find Hi-Q components precise, dependable and long lived. Send for samples and complete information.



Hi-Q Ceramic Capacitors are of titanium dioxide (for temperature compensating types) and are tested for physical dimen-sions, temperature co-efficient, power factor and dielectric strength. CI type with axial leads; CN type with parallel leads.



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Hi-O Choke Coils are uniform in their high quality performance. Ruggedly con-structed for long service.





5-TUBE POSTWAR RECEIVER

(See Front Cover)

THE i-f and audio system of a Stromberg-Carlson 5-tube postwar receiver, model 1000H, J and 1100H, is shown in the front-cover diagram this month.

The receiver features a loop tuned 14A7 r-f, resistance-coupled 14Q7 converter and 14A7 i-f stage. The latter stage has an unbypassed 150-ohm cathode resistor supplementing an avc bias.

Multiple Use of Diode

One of the diodes of a 14B6 used as a demodulator—avc—audio amplifier serves as a detector, the other being tied to the avc bus to supply a small bias from contact potential. The avc circuit is taken off at the high end of the detector transformer through a 2.2-megohm resistor, while the demodulated signal feeding the audio amplifier is taken from the low end through a 47,000-ohm resistor. This resistor, between detector and volume control, prevents overloading of the detector by the 14B6 triode, and with a 100 mmfd r-f bypass across a 1-megohm volume control serves to filter out r-f modulation.

A volume-control pickoff feeds the first audio grid through a .005-mfd capacitor and 10-megohm grid leak The low side of the control is bypassed to both the avc bus and B+through a pair of .05-mfd capacitors. Coupling to the power tube is achieved by a .01-mfd capacitor in series with a 120-ohm resistor, with a .47-megohm leak and 150-ohm cathode bias resistor, without a bypass capacitor. A .002-mfd capacitor across the first audio output shunts a fixed 200-mmfd bypass for increased bass effect. High frequency degeneration, or negative feedback, is provided by a 50-mmfd capacitor connected from the power tube's plate to the first audio plate. The capacitor is too small to affect bass frequencies appreciably; hence the bass gets through in full strength, the treble being somewhat attenuated as well as somewhat improved.







Simplicity of operation provides for the fastest settings ever developed for practical tube testing. Gives individual control of each tube element.

The New Speed Chek Tube Tester

MORE FLEXIBLE • FAR FASTER • MORE ACCURATE Three-position lever switching makes this sensational new model one of the most flexible and speediest of all tube testers. Its multi-purpose

New SQUARE LINE series metal case 10" x 10" x 5½", striking twotone hammered baked-on enamel linish. Detachable cover. Tube chart 8" x 9" with the simple settings marked in large easy to read type. Attractively priced. Write for details.

Model 2413

is another member of the NEW TRIPLETT Square Line

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Additional Features

- Authoritative tests for tube value; shorts, open elements, and transconductance (mutual conductance) comparison for matching tubes.
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- Sockets: One only each kind re-
- quired socket plus one spare.
 Distinctive appearance makes impressive counter tester.

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JOTS AND FLASHES

NOMPETENT Service Men will play an important role in keeping Mr. and Mrs. Public sold on television, said Paul E. Carlson, merchandising manager of the consumer products division of DuMont, at a recent meeting of home furnishings specialists. . . . About 60% of the present set owners plan immediate purchase of receivers, according to Frank Mansfield of Sylvania, who recently completed a set survey. Over 18,000,000 receivers are expected to be sold, he said. ... Lt. Kenneth C. Prince, U.S.N.R., was recently awarded the Bronze Star Medal by Vice Admiral Charles Lockwood, Commander Submarine Force Pacific Fleet. Lt. Prince was formerly legal counsel for the Sales Managers Club, Western group. . . . Marshall Field and Company, Chicago, has received an exclusive dealer franchise for Scott receivers. . . J. Kelly Johnson, formerly with Hammarlund Manufacturing Company, has opened a consulting engineering office at 55 West 42 Street, New York City. . . . Milo Radio and Electronics Corporation have announced the opening of their offices at 200 Greenwich Street, New York City. . . . G. Porter Burgess is now regional manager in the south and southwest for the Galvin Manufacturing Corporation. . . . Captain Pierre H. Boucheron, U. S. N. R., who recently returned from active duty and joined Farnsworth as director of public relations, addressed a recent meeting of the Radio Club of America in New York City. . . . Snyder Manufacturing Company, Philadelphia, have opened a sales office in the 333 Building, Chicago. Dwight Nelson and Leo Gibrich will headquarter at the new office.... Robert M. Brotherson has been named National Union district manager for Michigan, Kentucky, southern Indiana including Indianapolis, and Ohio, except Mahoning, Columbiana, Jefferson and Trumbull counties. . . . Ray L. Hoeffler has been appointed Zenith district sales manager for Philadelphia, Baltimore, Washington, and Norfolk. Edgar F. Lindgren will be district manager for Atlanta, Charleston, Charlotte, Raleigh, Jacksonville, Roanoke, Knoxville, and Chattanooga. Robert I. MacClellan will be district manager for Cleveland, Columbus, Toledo and Pittsburgh. . . E. J. Dykstra has been named Bendix district manager for the Chicago area. Mr. Dykstra was formerly with Farnsworth.

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undone to make the National Union family of products the most profitable of all for the service dealer. In engineering, in product performance, in sales policy, and in merchandis-

ing—the N. U. line is in a class by itself in meeting today's needs of the Service Dealer and Parts Distributor.

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