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WATFORD ELECTRONICS 35 CARDIFF ROAD, WATFORD, HERTS., ENGLAND MAIL ORDER, CALLERS WELCOME. Tel. Watford 40588/9	TTL 74 74126 32 74LS47 63174LS365 65 4078 21 CA3130 85 7400 11 74128 60 74LS47 63174LS365 65 4081 20 CA3130 85 7400 11 74128 60 74LS51 24/74LS365 65 4081 20 CA3130 45 7401 11 74132 41 74LS54 24/74LS367 65 4082 21 ICL7105E 795 7402 11 74136 55 74LS54 30/74LS376 65 4082 21 ICL7107 975 7403 12 74141 45 74LS43 150/74LS375 160 4086 73 ICM7205 1159 7404 12 74141 65 74LS47 4174LS375 160 4089 150 ICM7205 1159 7405 13 74143 250 74LS474 4174LS378 155 4093 159
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Saturday 9.00 am-6.00 pm. Ample Free Car Parking space available. POLYESTER CAPACITORS: Axial lead type (Values are In μF) 400V: 1n5, 2n2, 330n, 470, 648, 10m, 5n 39; 181 109; 22n, 33n 119; 47n, 68n 149; 100n 179; 150v, 320v, 24p; 330n, 470n 419; 680n 52p; 14F 64p; 24, 82p. 160V: 33v, 470n 419; 680n 52p; 14F 64p; 24, 82p. 160V: 33v, 470n 419; 680n 52p; 14F 64p; 24, 82p. 160V: 39v, 100n, 150n, 220n 119; 150n 109; 680n, 14P; 22p; 145, 242 32p; 4µ7 36p. 1000V: 10nF, 15n, 20p; 22n 22p; 47n 26p; 100n 38p; 470n 53p; 1µF 175p. POLYESTER RADIAL LEAD CAPACITORS (360V) 10nF, 15n, 20p; 32n, 27n, 88n, 100n 7p; 150n 109; 220n, 330n 13p; 470n 17p; 680n 18p; 1µF 22p; 1µ5 30p; 2µ2 34p. ELECTROLYTIC CAPACITORS: Axial lead type (Values are In µF) 500V 10 409; 47 68p; 50V 50, 100, 220 32p; 470 32p; 1000 50p; 40V: 22, 33, 8p; 100 12p; 220n, 3300 15p; 4700 98p; 50V 50, 100, 32p; 100 30p; 1000 40p; 40V: 22, 33, 8p; 100 12p; 2200, 3300 85p; 4700 98p; 50V 50, 100, 32p; 100 33p; 1000 40p; 40V: 22, 33, 8p; 100 12p; 2200, 3300 85p; 4700 98p; 50V 50, 100, 32p; 103 03p; 100 40p; 247 68p; 524, 100, 150, 100; 100; 100; 100; 100; 100; 100; 10	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$
329: 100. 330; 150. 330; 140: 329: 1000 480; 249: 100, 102, 400 380; 160: 400, 400 487, 200 487, 2100, 420 487, 420 487	7445 52 74179 53 74LS156 96 4025 19 4490V 525 Micital 779 7445 43 74180 45 74180 4501 19 MC3300P 120 7446 45 74181 110 74LS158 96 4027 45 4501 19 MC3300P 120 7447 45 74181 105 74LS161 98 4028 81 4503 69 MC3401 52 7447 45 74184 105 74LS161 98 4029 99 4506 51 MC3403 135 7450 15 74188 175 74LS163 102 4031 205 4506 51 MC3403 635 7453 15 74198 475 74LS163 102 4031 145 4511 150 MK50398 635 7450 15 74193 45 74LS168 154
100V:0.001,0.002,0.001,0.01// 6p SLIDER POTENTIOMETER Suiter LED 36 0.11/F By 0.2 Ap 50V:0.47 120 0.25W log and linear values 60m 0.26W log and linear values 60m 0.07P1 120 0.11/F By 0.2 Ap 50V:0.47 120 0.25W log and linear values 60m 0.07P1 600 MINIATURE TYPE TRIMMERS 5K 0.500K 0.1 single gang 600 0.25W log and linear values 60m 0.07P1 600 2-5.45pf5.45pf. 5.405F, 6.30-F, 10-40pF 22p Self Stick Graduated Bezels 7 Seg Displays COMPRESSION TRIMMERS 3.400F, 10-80pF 30p; 25-190pF 33p 0.25W 100 Ω-33M Ω Horiz 100 0.25W 100 Ω-33M Ω Horiz 100 0.25W 100 Ω-33M Ω Horiz 100 0.25W 100 Ω-33M Ω Horiz 100 1004 C Ch 3'' 89 101/04 C Ch 3'' 89 0.25W 100 Ω-33M Ω Horiz 100 0.25W 100 Ω-33M Ω Horiz 100 1004 C Ch 3'' 89 101/04 C Ch 3'' 89 0.25W 100 Ω-35M Ω Horiz 100 0.25W 100 Ω-35M Ω Horiz 100 1004 C Ch 3'' 89 101/04 C Ch 3'' 89 0.25W 100 Ω-35M Ω Film, High FND357 120 101/04 C Ch 3'' 89 101/04 C Ch 3'' 89 101/04 C Ch 3'' 89 </td <td>74/5 23 74/97 80 74/5.11 101 4039 320 702 75 75 325 74/6 22 74/198 150 74/5.181 386 4040 105 709C 14 pi 705C 35 NE561 395 7480 36 7150 175 74/5.181 386 4040 105 709C 14 pi 705C 11 35 NE561 395 7480 36 75491 92 74/5.181 140 4043 94 733 125 NE565 120 7483 47 74.5191 140 4044 95 741C 74 NE565 120 7484 80 74.5191 140 4044 95 741C 74 NE565 120 7485 40 74.5191 140 4047 127 741C 75 150 NE567 170 7486 67 74.500 114<</td>	74/5 23 74/97 80 74/5.11 101 4039 320 702 75 75 325 74/6 22 74/198 150 74/5.181 386 4040 105 709C 14 pi 705C 35 NE561 395 7480 36 7150 175 74/5.181 386 4040 105 709C 14 pi 705C 11 35 NE561 395 7480 36 75491 92 74/5.181 140 4043 94 733 125 NE565 120 7483 47 74.5191 140 4044 95 741C 74 NE565 120 7484 80 74.5191 140 4044 95 741C 74 NE565 120 7485 40 74.5191 140 4047 127 741C 75 150 NE567 170 7486 67 74.500 114<
10pF to 1nF 8p; 1 5nF to 10nF 10p. SitLVER MICA (Values in pF) 3-3, 4-7, 6-5, 10, 12, 18, 22, 33, 47, 50, 68, 4-7, 6-5, 10, 12, 18, 22, 33, 47, 50, 68, 100, 120, 500, 120, 150, 180 11p each 200, 250, 270, 300, 330, 360, 390, 470, 1000, 1200, 1800, 2200 15p each 15p each 16p each	7494 40 74LS29 22 74LS241 231 4053 72 AY-1-5051 145 TB A;00F 70 7495 30 74LS10 20 74LS242 232 4053 110 AY-1-67216 195 TB A;61 250 7495 34 74LS11 22 74LS242 232 4055 110 AY-1-67216 195 TB A;61 250 90 7497 98 74LS12 23 74LS244 155 056 134 AY-5-122A 450 TB A;810 96 74100 75 74LS13 38 74LS247 196 4060 115 CA3011 110 TC A;965 120 74107 50 74LS247 196 4060 115 CA3011 110 TD A1028 310 74107 30 74LS251 314 4061 1425 CA3011 100 1022 573 74107 30 74LS251 324
VOLTAGE REGULATORS* Suide 230V Suide 230V <t< td=""><td>T4118 56 74L530 22 74L5273 244 4070 32 CA3048 214 UAA170 150 74119 89 74L5273 224 4070 132 CA3080E 65 UAA170 150 74120 79 74L523 39 74L5279 66 4072 21 CA3080E 65 UAA170 150 74120 79 74L5333 39 74L5209 66 4072 21 CA3080E 85 ZN424E 130 74122 22 74L533 39 74L5209 468 0475 23 CA3080E 215 ZN425E 415 74123 37 74L5323 468 0476 85 CA3090A Q 275 ZN1034E 400 74125 33 74L542 97 74L5324 404 077 40 CA3123E 150 X1040E 485 74125 33 74L542 97 71450 150</td></t<>	T4118 56 74L530 22 74L5273 244 4070 32 CA3048 214 UAA170 150 74119 89 74L5273 224 4070 132 CA3080E 65 UAA170 150 74120 79 74L523 39 74L5279 66 4072 21 CA3080E 65 UAA170 150 74120 79 74L5333 39 74L5209 66 4072 21 CA3080E 85 ZN424E 130 74122 22 74L533 39 74L5209 468 0475 23 CA3080E 215 ZN425E 415 74123 37 74L5323 468 0476 85 CA3090A Q 275 ZN1034E 400 74125 33 74L542 97 74L5324 404 077 40 CA3123E 150 X1040E 485 74125 33 74L542 97 71450 150
100m A T O92 Plastic Casing SV 78L05 30p 79L05 65p ROCKER: SPST on/off 10A 250V 30p 8V 78L05 30p 79L05 65p ROCKER: SPST on/off 10A 250V 30p 8V 78L05 30p 79L05 65p ROCKER: SPST on/off 10A 250V 30p 8V 78L2 30p	AC176 25 BC182 10 BF167 30 MPF103 38 TIP3055 48 2N3133 43 AC188 24 BC182 10 BF173 25 MPF103 38 TIP3055 48 2N3135 33 AC188 24 BC183 10 BF177 25 MPF105 40 TIS44 45 2N3250 30 ACY18 60 BC1831 10 BF177 25 MPF105 40 TIS44 45 2N3250 30 ACY18 60 BC1841 10 BF178 25 MPF306 40 TIS8A 31 2N3250 30 ACY18 60 BC1841 10 BF170 30 MPSA05 15 TIS80 20 2N3663 14 ACY20 53 BC1841 10 BF197 38 MPSA05 16 TIS91 2N3702 10 ACY22 60 BC2121 9
CAPACITORS 0 2 3550 F with BA102 A129 28 Range 2V7 to 39V 400mW Thyristors Dielectric 0 2 3550 F with BA102 15 39V 400mW 0:8A/100V 30p 0:8A/100V 30p 500pf 205 D Drive 325 p BY126 12 Range 2V7 to 39V 400mW 0:8A/100V 30p 0:8A/100V 30p 6 1 Ball Drive 00 208/176 255 p BY127 12 33V.1 13W 5A00V 33 5A800V 33 Jail Drive 00 208/176 250 p CR03 148 15p each 5A800V 35 5A800V 35 5A800V 35 5A800V 35 5A800V 35 Drum 54mm 40p 250 pF 163 p CA70 12 Z5J 16 8A800V 35 5A800V 35 Drum 54mm 40p 13 x 310p 550p CA79 15 DA17 12 A500V 35 5A800V 35 01:365pF 245p 100, 150pF 275p CA79 15 BRIDGE 12A300V 59 12A300V 59 02 285pF 285 p 00:3 x 25pF 339p CA85 14 BRIDGE 12A300V 182p	A Diego 12 B C307 12 B F224A 18 O C35 125 Z TX304 17 2 N3710 10 A F114 60 B C307 12 B F244A 18 O C35 125 Z TX304 17 2N3710 10 A F114 60 B C307 12 B F244B 24 O C36 130 Z TX314 24 2N3711 10 A F115 60 B C327 12 B F244B 24 O C41 152 Z TX314 24 2N3713 215 A F118 75 B C328 12 B F256 60 O C42 48 Z TX314 20 2N3711 195 A F130 70 B C441 27 B F256 80 O C42 48 Z TX501 15 2N3772 195 A F130 70 B C441 27 B F257 30 O C44 52 Z TX501 15 2N373 283 A F136 50
DENCO COILS RDT2 98 p OAst (plastic case) 2N4444 140 p 'DP' VALVETYPE RFC 5 chokes 98 p OAst 1A/50V 20 2N5064 33 Range 1 to 5 Bi., RFC 1(9mH) 102 p OA202 1A/100V 28 2N5064 33 Rd., YI. Wht. 92 p IFT 13; 14; 15; OA202 1A/100V 28 E1060 38 6-7 B.Y.R. 85 p 16; 17 92 p IN914 1A/400V 28 TIC44 28 1-5 Green 99 p IFT 18/45 109 p IN4001/2 5 1A/600V 34 TIC45 45 Rd., Wht. 1-05 p TOC 1 92 p IN4004/5 2A/30V 34 TIC45 45 B9A Valve Holder MWSFR 112 p IN4004/5 2A/100V 48 3A/100V 48 VEROBOARD 0.1 0.15 IS44 2 2A/400V 53 3A/200V 48 VEROBOARD 0.1 0.15 IS44 2 2A/400V 34/400V <td>BC109 11 BC559C 10 BFR79 24 OC81 36 2N699 30 2N3906 17 BC109B 12 BCY70 14 BFR80 24 OC82 50 2N706 19 2N4037 52 BC109C 12 BCY70 14 BFR80 24 OC83 48 2N706 19 2N4058 17 BC117 20 BCY71 14 BFR81 24 OC83 48 2N708 19 2N4058 17 BC117 20 BCY71 14 BFR81 24 OC83 48 2N708 19 2N4058 17 BC117 20 BCY72 16 BFX84 20 C140 100 2N330 18 2N4051 17 BC140 26 BD133 50 BFX85 28 OC171 45 2N1131 22 2N5136 42 BC142 26 BD135 30</td>	BC109 11 BC559C 10 BFR79 24 OC81 36 2N699 30 2N3906 17 BC109B 12 BCY70 14 BFR80 24 OC82 50 2N706 19 2N4037 52 BC109C 12 BCY70 14 BFR80 24 OC83 48 2N706 19 2N4058 17 BC117 20 BCY71 14 BFR81 24 OC83 48 2N708 19 2N4058 17 BC117 20 BCY71 14 BFR81 24 OC83 48 2N708 19 2N4058 17 BC117 20 BCY72 16 BFX84 20 C140 100 2N330 18 2N4051 17 BC140 26 BD133 50 BFX85 28 OC171 45 2N1131 22 2N5136 42 BC142 26 BD135 30
(copper clad) (plain) 3A/4000 16 4A/800 12 AA/800 12 21 × 31 65 p 38 p 24 p 3A/4000 27 6A/100 73 8A400 44 21 × 5 55 p 50 p 31 p 3A/100 V 36 6A/200 78 8A800 100 32 × 32 55 p 50 p - - 3A/100 V 36 6A/200 V 78 8A800 100 32 × 32 55 p 50 p - - - 6A/200 V 78 12A400 V 70 22 × 17 168 p 135 p 92 p - - 6A/400 V 85 12A400 V 73 32 × 17 218 p 180 p 120 p We stock a W18 DiL 50 16A100 V 95 42 × 17 218 p 180 p 120 p of Electronic - - Pkt of 35 pins 220 p 22 p of Electronic - 25A800 V 150 Plin insertion tool 144 p Magazines ST2 25 T28000 120	BC153 20 BD140 30 BFY52 21 TIP31 50 2N1671B 120 2N5457 32 BC154 13 BD145 175 BFY56 32 TIP31A 38 2N2160 350 2N5457 32 BC154 10 BD145 175 BFY56 32 TIP31C 50 2N2219A 22 2N5458 32 BC158 10 BD2055 110 BFY71 20 TIP31C 50 2N2219A 22 2N5458 32 BC158 11 BD222 75 BSX20 20 TIP32C 55 2N2220A 32 2N5455 35 BC169 28 B0378 70 BSY65 30 TIP33C 55 2N2426 12 3N128 112 BC168 10 B0517 70 BU105 115 TIP34C 63 2N2484 25 3N140 112 BC169 10 B05658

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Everyday Electronics, March 1980

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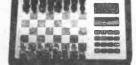
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Everyday Electronics, March 1980

Britain's first comp

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*Use a 600 mA at 9 V DC nominal unregulated mains adaptor. Available from Sinclair if desired (see coupon).

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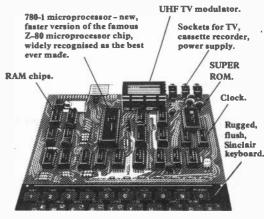
The unique Sinclair BASIC interpreter... offers remarkable programming advantages:

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- FOR/NEXT loops nested up 26.
- Variable names of any length.
- BASIC language also handles full Boolean arithmetic, conditional expressions, etc.
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- PEEK and POKE enable entry of machine code instructions, USR causes jump to a user's machine language sub-routine.

- High-resolution graphics with 22 standard graphic symbols.
- All characters printable in reverse under program control.
- Lines of unlimited length.

...and the Sinclair teach-yourself BASIC manual.

If the features of the Sinclair interpreter listed alongside mean little to you-don't worry. They're all explained in the specially-written 96-page book *free* with every kit! The book makes learning easy, exciting and enjoyable, and represents a complete course in BASIC programming-from first principles to complex programs. (Available separately - purchase price refunded if you buy a ZX80 later.)



Everyday Electronics, March 1980

Science of Cambridge Ltd

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Projects...Theory...

and Popular Features ...

The words home improvement will have an ominous ring about them for many a householder. They conjure up visions of all those jobs that (we are likely to be told) should occupy quite a lot of our spare time.

Home improvement can imply all kinds of jobs from routine redecorating to grandiose schemes involving removal of entire interior walls and other structural changes. Yet they have one thing in common. Any of these operations can bring us up against the problem of buried pipes or mains wiring. For not always is the precise course of these vital service arteries clearly charted. Indeed, in some circumstances there can be a hazard even in the simple and apparently harmless task of knocking a nail or screw into a wall to hang a picture.

The Cable and Pipe Locator will therefore be a sensible addition to the handyman's tool kit. Since this tool will be used only occasionally, it has been designed to be used with a small portable radio. In this way the locator unit is kept simple and uses only a few components. Good conservation of resources, please note.

An ever-useful device for the kitchen is an electronic timer. The design presented this month has an extensive range, switchable in 1 minute intervals from a minute to 2 hours. Build this and you won't be accused any more of "wasting your time on that electronics nonsense instead of getting on with the wallpapering." Well, maybe.

Moving on now to other parts of the house, harmony in the home often hangs on disciplined use of the hi-fi by the music fanatic whose choice of music or volume level may not be shared by other members of the family. Listening via headphones is an obvious answer.

Headphones are also convenient for the person living alone, or wherever space limitations do not warrant the outlay on loudspeaker enclosures or a high power amplifier. The Stereo Headphone Amplifier will meet requirements such as these very economically.

The projects mentioned so far have well-defined functions to perform in the home. Coming to the *Doorbell Register* the situation is somewhat different. Just how useful this little device will prove to be it is hard to foretell.

Despite its name, its applications obviously are not confined to the front door. It could for example be employed to provide evidence that a room had been entered, or an article disturbed...food for thought at anyrate.

feel Semmet

Our April Issue will be published on Friday, March 21. See page 175 for details.



Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

We cannot undertake to engage in discussions on the telephone.

Component Supplies

Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

All reasonable precautions are taken to ensure that the advice and data given to readers are reliable. We cannot however guarantee it, and we cannot accept legal responsibility for it. Prices quoted are those current as we go to press.



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MARCH 1980

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Back Issues

Certain back issues of EVERYDAY ELECTRONICS are available worldwide price 70p inclusive of postage and packing per copy. Enquiries with remittance should be sent to Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 0PF. In the event of non-availability remittances will be returned. * Not available: October 1978 to May 1979.

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By G. Hallam

T HIS project arose from the alarming and often dangerous activity of drilling holes in walls in which mains wiring and pipes might be buried.

A simple device was required to give an audible indication of the position of wires and metal conduits with sufficient sensitivity and positional accuracy to avoid striking them. The subsequent device featured in this article has proved invaluable to the author and every handyman ought to benefit from the confidence this device will bring him.

Since copper wire, metal pipes and steel conduits have to be detected, the device is effectively a metal locator and can be used as such in the garden and field and on the beach to find coins and other buried objects of historical interest.

In order to keep construction simple, the cable detector makes use of an ordinary portable transistor radio having a medium or long wave band. The detector generates radio waves, the frequency of which is changed when metal is brought near to the search coil.

- POST OFFICE LICENCE -

Since this pipe locator actually radiates a radio frequency signal, it requires a Pipe Finder Licence. This can be obtained from the following address: Home Office Radio Regulatory Department, Waterloo Bridge House, Waterloo Road, London, SE1 8UA. At the time of going to press this cost $\pounds1.40$ for 5 years although readers should check with the Home Office at the above address for the current position. These waves interfere with the internally generated intermediate frequencies in the radio to produce an audible tone from the radio. This "beat" frequency varies in tone as metal passes across the search coil.

THE CIRCUIT

The circuit shown in Fig. 1 is a weak radio transmitter and as such needs to be licenced (see licensing address below). The circuit is essentially a Colpitt's oscillator. This consists of a tuned circuit made up of L1, C1 and C2 which forms the collector load of the transistor TR1.

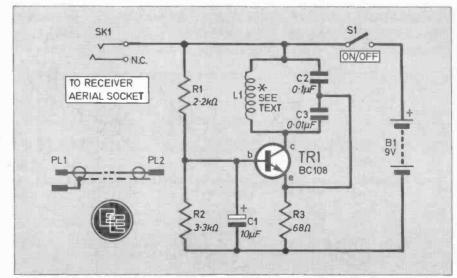
When switched on, the circuit oscillates at a frequency determined by the inductance of L1 and the values of Cl and C2. Because it is necessary to use a portable radio as a detector, r.f. harmonics are not filtered out.

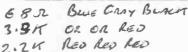
The search coil consists of 100 turns of 33 s.w.g. enamelled copper wire wound on a 65mm diameter former.

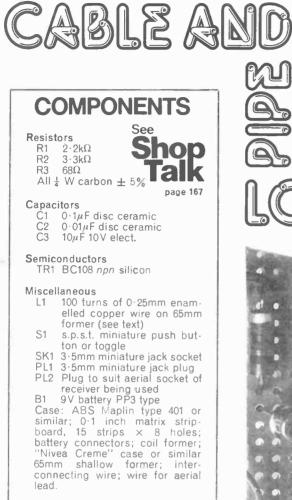
RADIO RECEIVER

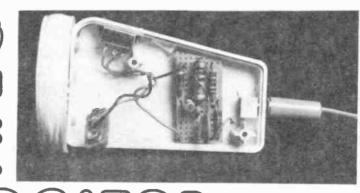
Radiations from this circuit pass to a nearby radio receiver and produce the audible beat notes. The effectiveness of the unit is improved if a flying lead is connected between the jack socket of the detector and the aerial socket of the receiver, although just winding this flying lead round the radio may be as effective.

Fig. 1. Circuit diagram of Cable and Pipe Locator.









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(above) The completed unit with one side panel removed showing positioning of the circuit board, on/off switch and jack socket.

(left) The finished component board.



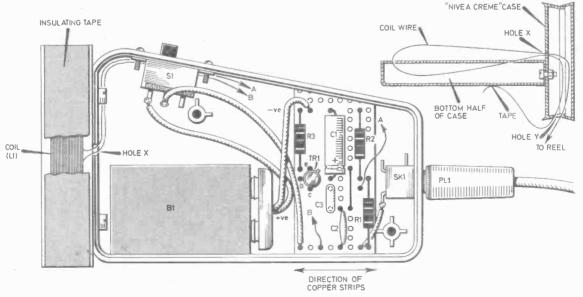
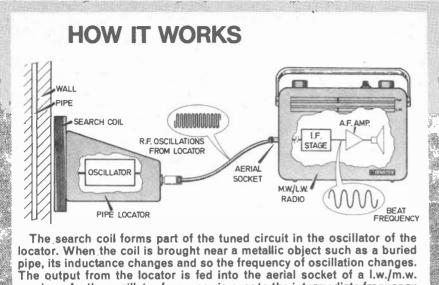


Fig. 2. Interior view of the unit showing component board layout and inter-wiring. Note that there are no breaks in the copper strips on the circuit board. Inset above right shows details of coil construction.



pipe, its inductance changes and so the frequency of oscillation changes. The output from the locator is fed into the aerial socket of a l.w./m.w. receiver. As the oscillator frequency is near to the intermediate frequency of the receiver, a beat note is produced in the loudspeaker. Any change in the oscillator frequency caused for example by the coil being near a pipe will cause the beat frequency to change and so the presence of the pipe will be shown up by the change in the note from the loudspeaker.



CIRCUIT BOARD

The components are mounted onto a small piece of 0.1 matrix stripboard 15 strips by 8 holes and shaped so that it fits into the case in the position shown in Fig. 2.

The stripboard, PP3 battery Bl, jack socket SK1, and switch S1, are housed in a hand-held ABS case with the coil former attached to the wide end as shown in Fig. 2. The holes for the jack socket and switch are made in the bottom half of the case in the positions indicated.

The coil former is made from a 65mm shallow former such as a small size "Nivea Creme" case assembly of which is shown in Fig. 2. The wide end of the case is filed flat and the "Nivea" case fixed to the bottom half with two 6BA nylon screws and nuts so that when the two halves of the ABS case are assembled, the coil former is central.

Two small holes are drilled at points X and Y (2mm) through which the leads to the coil former are fed.

COIL

The free end of a reel of 0.25mm enamelled copper wire (approximately 20 metres are required) is fed into the case via holes Y and X in the case of the coil former. A small loop of wire is left inside the case and the end brought out via X and Y and temporarily attached to the case with Sellotape (see Fig. 2). The lid of the coil former is glued to its other half with cyanoacrylate or other suitable adhesive ("Superglue") and when this has set 100 turns of copper wire are wound round the coil former. The end of the coil winding is then cut off and soldered to the free end of the wire which was temporarily taped to the case.

The combination is then pulled into the ABS case via holes Y and X and the coil leads cut to length and soldered to the stripboard. The coil windings can be protected by applying a few turns of PVC tape.

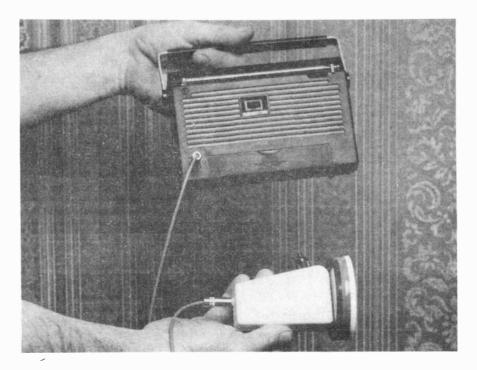
The jack socket SKI and switch SI are now mounted in their holes and the circuit connections completed. The detector is now ready for use.

SETTING UP

The detector and the radio receiver are switched on and the tuning dial moved until a whistle is heard as you tune across a station. Select a station where the whistle is loudest and the actual transmission quietest. The note should change in pitch when the coil of the detector is passed over various metallic objects.

Ferrous and non-ferrous objects will be found to cause the note to change in different ways allowing the user to distinguish between copper and iron pipes. This makes the detector an ideal aid to the handyman as it can be used to trace the path of electric cables and conduit buried in walls or for tracing water and heating pipes under floors.

The completed Cable and Pipe Locator plugged in to a radio receiver and ready for use.



Everyday Electronics, March 1980



By Dave Barrington

Screwdrivers

One of the many problems we encounter in our work is the "fun and games" one experiences trying to fix screws in awkward places when finally putting projects together. Also, we are sometimes asked by beginners "Can you recommend what size screwdriver I will need when I make up my basic tool kit?"

Well, the answer we usually give is: "As many as your pocket will allow making sure they are of varying lengths and tip sizes, including at least two Phillips, two Pozidriv types and one standard blade stubby type".

A good range of tip sizes would be $2 \cdot 5$, 3, $4 \cdot 8$, $5 \cdot 6$ mm and possibly the larger 7mm and $9 \cdot 5$ mm for heavy duty work. Of course, each individual will have his own ideas and the final choice is yours. Always go for the best quality you can afford.

The reason for suggesting a stubby type is because we have found it a good general purpose screwdriver and invaluable in the workshop for those difficult corners. Its surprising how much leverage you can exert with these types.

Just released on to the market is a new stubby screwdriver from J. Stead & Co. Ltd., and features ratchet operation. Known as the Steadfast Screwmaster it has a 8mm blade and measures only 90mm, including the handle. The ratchet mechanism is a scaled

The ratchet mechanism is a scaled down adaption of the clutch type and, when in use, as long as the blade has the resistance of the screw to hold it the required operation is gained by rotating a selection ring. This, the makers claim, enables one hand operation.

The price of the Steadfast Screwmaster is £2.25 and details of nearest stockists can be obtained from J. Stead & Co. Ltd., Dept EE, Greenland Road, Sheffield S9 5BW.

Catalogues

Four catalogues have been received this month.

The latest Greenweld 1980 Components catalogue contains 48 pages and is well illustrated throughout. Looking through the pages it is noticeable that many items have actually been reduced in price from last year's catalogue.

The catalogue cost 40p plus 20p postage, but it contains five 12p discount vouchers for use on the first order over £3; thereafter on orders over £2.

Two new catalogues have been issued by Tandy this month. They are the Tandy 79/80 Electronics Catalogue and the Tandy TRS-80 Microcomputer Catalogue.

With the rapid expansion of home computing, Tandy's have found it necessary to produce a 24 page special for their now famous TRS-80 computer. This catalogue lists the hardware and software presently available.

The new 116 page Electronics Catalogue contains many first time items and the components section seems to increase every year and is well worth obtaining.

It is a pleasure to bring to the attention of readers, schools, colleges, universities and even industry the Toolrange 79/80 catalogue.

Making full use of colour, this 104 page catalogue is well presented and contains one of the most comprehensive stocks of workshop tools and complete transportable kits we have come across. For anyone just about to replenish his workshop we strongly recommend they obtain a copy of this catalogue before making their final choice.

The sections on pliers, cutters and screwdrivers will take some beating and the range of soldering irons and soldering equipment is quite impressive.

For more details readers should contact Toolrange Ltd., Upton Road, Reading, RG3 4JA.



Steadfast Screwmaster

Mail Order

Many of our readers will, no doubt, remember the excellent service provided by Doram Electronics Ltd., and will welcome the news of its relaunch under new ownership.

Originally formed by the Electrocomponents Group to distribute RS Components' components by mail order, it ventured into the complete kit and readymade product market with only mild success and ceased trading as part of a group reorganisation.

The new owners are the Dutch De Boer Group, one of Europe's largest component mail order suppliers, who will administer the Doram mail order business from a UK based distribution centre. The product range, which will include both kits and components, will be based on the current Dutch Components Catalogue.

For further details write to Doram Electronics Ltd., Dept EE, Fitzroy House, Market Place, Swaffham, Norfolk PE37 7QH.

If the service provided by the new company measures up to the old, then readers will be on to another winner.

Change of Address

We have just been informed by Home Radio (Components) Ltd, that due to refusal of planning permission they have had to move to new premises.

The new address for callers is: 269a, Haydons Road, Wimbledon, London, SW19 8TY.

The address for Mail Orders is: PO Box 92, 215 London Road, Mitcham, Surrey, CR4 3HD.

CONSTRUCTIONAL PROJECTS.

Radio Control System

Some good news for readers who may be worried about their ability to make up the p.c.b.s for the *EE Radio Control System*. We understand that Proto Design is already producing the receiver and transmitter boards and are planning to produce a complete kit of boards.

Made from fibreglass, the boards are supplied pre-drilled and "roller tinned" (for ease of soldering) and cost $\pounds 1.47$ for the transmitter and 88p for the receiver. A charge of 25p is levied to cover post and packing.

For further information readers should contact Proto Design at 14 Downham Road, Ramsden Heath, Billericay, Essex, CM111PU.

Any special items for the Radio Control System are fully covered in the article.

Stereo Headphone Amplifier

The LM301 AN called for in the *Stereo Headphone Amplifier* is a low noise plastic dual-in-line version and may prove to be troublesome, but should be available from Marshall's or Watford Electronics and is certainly listed in Electrovalue's catalogue. The LM301 AH may be used but being a metal can type the leads will have to be carefully pre-formed. Note that the negative supply, pin 4, is connected to the metal case.

The case used in the prototype was one of the common metal chassis with a base plate fixed to the four corner strengtheners.

Touch Switch

Some difficulty may be experienced with the Darlington transistors called for in the Uniboard-Touch Switch project.

The MPSA14 is available from Maplin or Marshall's and alternative devices are the MPSA12, MPSA13, both available from Watford or Marshall's.

Cable and Pipe Locator

The Vero plastics hand-held control box used to house the circuit board for the *Cable and Pipe Locator* is now stocked by a number of our advertisers and should present no problems. Apart from the above the rest of the components should be readily available.

All components for the *Doorbell Register*, 5—*Range Current Limiter* and the *Kitchen Timer* are fairly common and should not be difficult to obtain.

The loudspeaker called for in the Timer can be any type from 40 to 80 ohms.

Stereo Headphone Amplifier

Π

By R. A. Penfold

WHEN people first try out a pair of stereo headphones they are usually surprised by the high quality of reproduction. Even inexpensive stereo headphones (which usually have ordinary miniature loudspeakers as the transducers) seem to provide quite a high level of fidelity, the main shortcoming being a lack of bass and treble response which is easily compensated for by adjustment of the tone controls. Some of the higher quality types are still very inexpensive by loudspeaker standards, and offer true high fidelity reproduction.

Although headphones are normally considered to be an adjunct to a hi fi system, where low cost is of prime importance there is no reason why they should not be used in place of speakers; indeed, this would seem to be an eminently sensible approach. There are other advantages to such a system, such as the fact that a system of this type is perfectly well suited to use where the amount of available space is severely restricted. The absence of large and heavy loudspeakers results in ease of portability.

Also, a great advantage of headphones is that even when listening at high volume levels, late at night, there will be no annoyance caused to others.

This article describes a simple and inexpensive amplifier which is specifically intended for use with stereo headphones. The prototype was used in conjunction with a pair of Sennheiser HD400 phones and offers a quality of reproduction that would be difficult to better for the monetary outlay involved.

Of course, the unit will operate well with any normal type of stereo

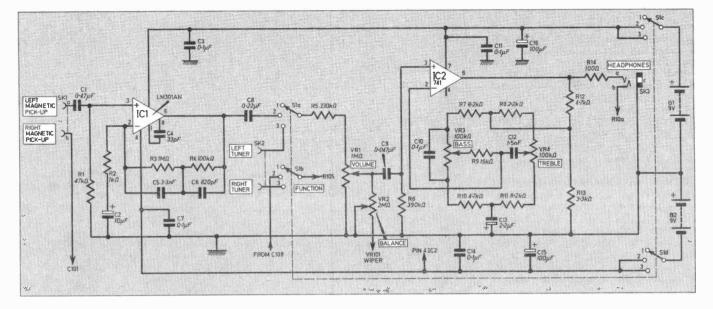
Fig. 1. The circuit diagram of the Stereo Headphone Amplifier.

'phone that does not require a high level output (i.e. not a special type that requires direct connection to the loudspeaker output of an amplifier).

The unit can be employed with a magnetic, crystal or ceramic pick-up, tape deck, or tuner. The amplifier is self contained as *it* is powered from a couple of PP3 9 volt batteries. The total current consumption of the amplifier is about 5mA, equally shared between main and preamp.

CIRCUIT DESCRIPTION

The circuit diagram of the Stereo Headphone Amplifier appears in Fig. 1. This only shows the circuit for one channel of the amplifier, but the other channel is essentially the same. Supply decoupling capacitors, S1, SK3 and balance control VR2 are common to both channels, but the remaining circuitry is duplicated in the other channel. In the constructional drawings and components lists the components in the other channel are given the same identification numbers, but preceeded by "100".



MAGNET PREAMPLIFIER

The main amplifier requires an input level of about 200mV r.m.s. or more in order to provide a good output volume level. A magnetic cartridge provides a much lower output than this; usually something in the region of 3 to 10mV r.m.s.

The audio signal applied to a record is given a degree of treble boost and bass cut. The treble boost results in an improved signal to noise ratio as it enables a flat frequency response to be obtained at playback with a certain amount of treble cut applied to the signal. The bass cut is needed in order to prevent groove wall collapse at high bass signal levels. A suitable amount of bass boost must be applied at playback in order to compensate for the original bass cut.

Ceramic and crystal cartridges have an output characteristic which automatically provides the necessary equalisation, but this is not the case with magnetic pick-ups. Therefore the preamplifier has to provide both gain and the appropriate equalisation.

The circuit of the magnetic pickup preamplifier employs an LM301AN operational amplifier which provides low levels of noise and distortion.

Like the main amplifier, this circuit is of the non-inverting type. The input signal is applied to the non-inverting input of ICl via Cl, and Rl biases this input to earth potential. This resistor also sets the input impedance of the circuit at about 47 kilohms, which is the load impedance or required by most magnetic cartridges.

RIAA EQUALISATION

Resistors R3 and R4 bias the inverting input from the output of the amplifier, and together with R2, C5 and C6 they form a negative feedback network which sets the gain of the circuit at a suitable level and provides the RIAA equalisation. It is the frequency selective feedback introduced by C5 and C6 which gives the equalisation bass boost and treble cut. At high frequencies the impedance of C6 falls in relation to that of R4, producing the treble cut. At bass frequencies the rising impedance of C5 produces reduced feedback and hence the required bass boost as well.

In a theoretically perfect circuit C2 would not be needed and R2 connected direct to earth. In a practical circuit this would result in the amplifier having a very high d.c. gain, with the quiescent output voltage tending to drift well away from earth potential in consequence. This would result in clipping of the output signal. Compensation capacitor C4 prevents IC1 from becoming unstable; C3 and C7 are supply decoupling capacitors and are common to both channels. All the other components are duplicated in the other channel.

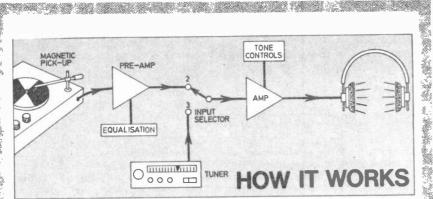
MAIN AMPLIFIER

The circuit is based on a conventional non-inverting operational amplifier circuit utilising the well known 741C device. This type of amplifier requires dual (positive and negative) balanced supplies with a central 0V earth rail. This is supplied by a couple of PP3 batteries, one to supply the positive rail and the other to supply the negative potential. A mains power supply could be used, but this would be more complicated and would make it more difficult to obtain a really low background noise level. Background hum and noise tends to be very noticeable when using headphones. Two poles of S1 provide on/off switching.

Resistor R6 biases the non-inverting (+) input of IC2 to earth potential, and the inverting (-) input is biased from the output of IC2 via the negative feedback loop which consists of R12 and R13.

The amount of feedback applied to the circuit determines the voltage gain of the amplifier, minimum feedback corresponding to maximum voltage gain.

Destat	COMPON		R°UE
Resistors R1,101 R2,102 R3,103 R4,104 R5,105 R6,106 R7,107 All 1 W c	330kΩ 390kΩ	R8,108 R9,109 R10,110 R11,111 R12,112 R13,113 R14,114	8·2kΩ 4·7kΩ
VR2 VR3,103	eters 1MΩ log. law carbon 2MΩ lin. law carbon 100kΩ lin. law carbon 100kΩ lin. law carbon	dual-gang (1 off)	
C7 C8,108 C9,109 C10,C110 C11 C12,112	0.047μ F polyester typ 10μ F 10V elect. 0.1μ F polyester type 33μ F polyestyrene 3.3 nF polyestyrene 820μ F polyestyrene 0.1μ F polyester type 0.22μ F polyester type	C280 C280 e C280 e C280 C280 C280 C280	See Shop Talk page 167
Semicondu IC1,101 IC2, 102	i ctors LM301 AN low noise c 741 op-amp 8-pin d.i.l	op-amp (8-pin d.i.i I.	.)
Miscellane S1 SK1,2 SK3 B1,B2	ous 4-pole 3-way rotary sv twin phono sockets (2 standard stereo jack : 9V type PP3 (2 off)	2 off)	
battery con	: 0·1 inch matrix size: nectors for B1,B2 (2 pa case, approximate din	ir): control knobs	holes, 17 strips × 25 holes; (5 off); 6B A fixings and solder 50 × 60mm—see text.
compon to their Thus tw remaind	ents in identical positi circuit reference as car 10 of each value are r	ions in the other n be seen above a equired except fo	e channel are shown. The channel have "100" added and in the layout diagrams. or the potentiometers. The th channels, and of course



Headphones require a driving power of only a few milliwatts to provide realistic volume levels. The unit does not, therefore, incorporate power amplifiers, but drives the phones from low level amplifiers using a single operational amplifier i.c. in each channel. The gain of this circuit is controlled by a negative feedback loop, with the usual bass and treble tone controls are incorporated.

Two sets of input sockets are provided, and a switch selects the desired input. The unit can be used with all normal types of hi fi gear, including magnetic cartridge. This requires extra gain and frequency response shaping (equalisation) which is provided by an optional preamplifier.

TONE CONTROLS

The gain of the circuit is not solely determined by R12 and R13 since a simple tone control network is interposed between their output and the inverting input of the I.C.I. Potentiometer VR4 is the treble control, and when the slider of this potentiometer is at or near the top of its track, C12 will introduce a comparatively large amount of feedback at treble frequencies. This results in treble cut being applied to the circuit.

With VR4 slider towards the other end of its track, C12 will tend to decouple the high frequency feedback causing reduced feedback at these frequencies and a consequent boost in treble response. The treble response will be flat with VR4 slider at roughly the centre of its track.

Potentiometer VR3 is the bass control, and in conjunction with C10 this can increase or decrease the relative amount of feedback at bass frequencies, thus providing bass out and boost respectively. Resistor R9 minimises interaction between the two tone control networks.

Two sets of input sockets are provided, the desired input being selected by means of S1. From here one pair of input signals is fed to the input of the amplifier through R5, volume control VR1, and input d.c. blocking capacitor C9. The other pair of inputs are fed straight to the magnetic pick-up preamplifier whose output goes to S1 position 2. Potentiometer VR2 forms a conventional balance control in conjunction with R5. The output to the headphone is obtained by way of series resistor R14. When using 8 ohm impedance headphones this resistor introduces considerable losses, but this is necessary since the output amplitude of the unit will be far too high for this type of headphone; R14 also reduces the loading on the amplifier to an acceptable level when using low impedance headphones. Many high quality phones have a fairly high impedance, usually in the range 100 to 2,000 ohms; R9 then provides little or no significant attenuation, but this produces good results since higher impedance headphones require higher drive voltages.



CASE

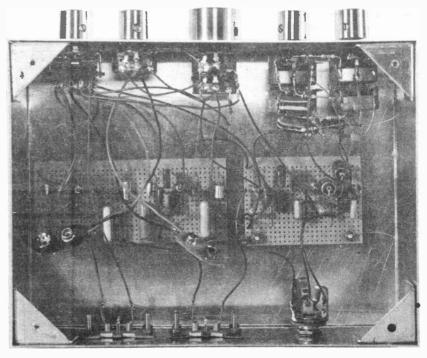
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A suitable housing for the project consists of a $203 \times 152 \times 63$ mm aluminium chassis with baseplate. The baseplate is fixed to the corner pieces of the chassis using four 6BA self tapping screws which are not normally supplied with the chassis or baseplate. The mounting holes in the corner pieces are drilled using a No. 42 or similar twist drill, and the holes in the baseplate are made using a No. 31 drill, or an equivalent.

FRONT PANEL

One of the 203 x 63mm sides of the case is used as the front panel, with VR1 centrally on this. The tone controls, VR3 and VR4 are mounted to the right of this with VR2 and S1 mounted symmetrically opposite these on the left (see photographs).

Plan view of the completed prototype amplifier.



Stereo Headphone Amplifier

SK1 (LEFT)

1 2

000000

C C c BA

BA BB BD BE BF BG BH BJ BJ

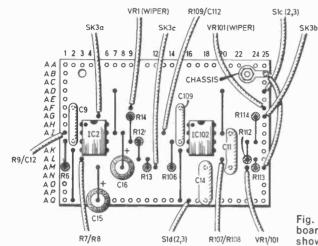
BL BA BN 80 ò

8/

7 8 9 10

0 0 0 0000

12 14



25	24		22	2	20		18		16		14		12		10	9	8	7	6	5	4	3	2	1
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0	0	C	0	0	0	0	0	0	0	Ô	0	0	0	0	0	0	0	0	0	Ó	0	0	0	0
0	0	Ó	0	Ô	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	0	0	0	0		0	0	0	0	0	0	0	0	0	0	0	0		0	0	.0	0	0	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	Ó	0	0	0	0	0	0
1	_	0	0				Q												0	0	0	0		0
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÷.,	0	0	0	0	0	0	0	0		0	0		-	0	_	0	0	0	0	0	Ó	0	0	0
0	0	0	0	0	0	0	0	0	0	0	0	0	¢		0	0	0	0	0	0	0	0	÷.	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	O'
0	0	0	0	Ó	0	0		0			Q	0	0	Ó	0	Ó	0	0	0		0	0		0

Fig. 2a. The layout of the components on the topside of * board A and the breaks to be made on the underside. Also shows the positions of the flying leads to reach other components. Note the connection to chassis via solder tag/mounting bolt.

C101

00000000

SKI (RIGHT)

R10

28

00 0000

32

34

0000

C108

30

S1b(2)

40 41

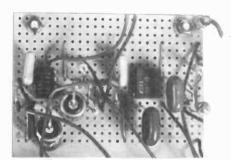
S1c (2,3)

S1a(2)

20 22

C8

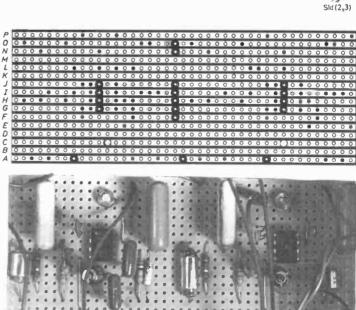
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The completed prototype main amplifier board.

Fig. 2b. The magnetic preamplifier board. Shows the $\stackrel{O}{}_{N}$ layout of the components on the topside of the board, $\stackrel{N}{M}$ breaks to be made in the copper strips on the underside, Land flying lead connections. A connection to chassis is K_{L} necessary via solder tag and board mount.





The completed prototype preamplifier board.

Stereo Headphone Amplifier



Layout of the controls and suggested labelling on the front panel;

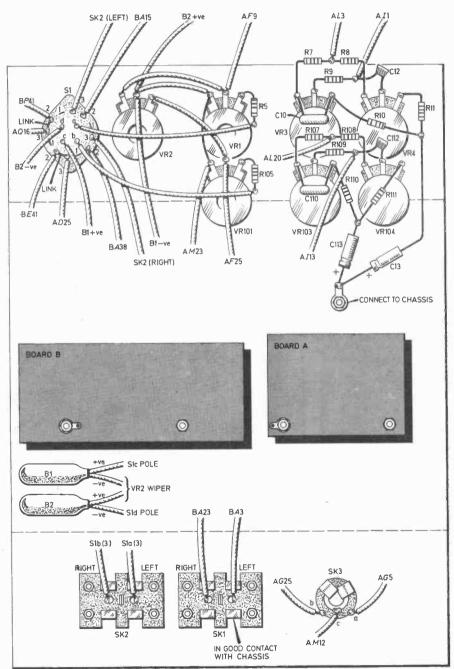
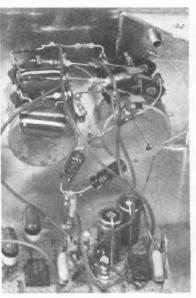
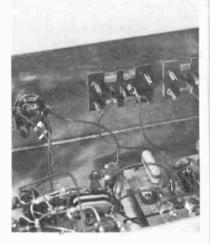


Fig. 3. Complete interwiring details. For clarity the front and rear panels have been laid down. Double-sided adhesive tape can be used to hold the batteries in position.



Close-up view of the interwiring of the tone controls.



The wiring to and from the sockets fitted on the back panel.

All these controls require standard 10mm diameter mounting holes. In the prototype a silver card was cut to size and lettered using Letraset, to improve the appearance.

The input and output sockets are mounted on the rear panel, and two twin phono sockets are used as the input sockets on the prototype. However, DIN or any other audio type connectors can be used here if preferred. The output socket is an ordinary open construction type standard stereo jack socket, and requires a 10mm diameter mounting hole.

COMPONENT PANELS

The circuitry is assembled on two boards, one holding the main amplifier and the second the magnetic pick-up preamplifier. The latter may be omitted if not required and the connections from S1 taken to SK1 to form a second identical stereo input.

Some of the components of the main amplifier board, board A, are wired up on a 0.1in matrix stripboard having 17 copper strips by 25 holes see Fig. 2a. Commence construction of this by cutting out a panel of the appropriate size using a hacksaw. Then filed down any rough edges of the board and drill the two mounting holes for 6BA or M3 clearance (about a No. 31 or 3.2mm diameter twist drill). Then the 15 breaks in the copper strips are made using either the special spot face cutter tool or a small hand held twist drill.

Next the components and link wires are soldered into position, the

two i.c.s being left until last. The latter are very inexpensive types, and the use of i.c. sockets is probably not justified provided reasonable care not to overheat these components is taken when soldering them.

The completed component panel is mounted on the underside of the top panel of the case just to the rear of VR3 and VR4. Spacers must be used over the mounting bolts to hold the underside of the panel clear of the metal case. The component panel is provided with a chassis connection via a soldertag mounted on one of the mounting bolts for the panel.

Note that the panel cannot be finally bolted into position until it has been wired up to the rest of the unit.

CONTROL WIRING

In order to minimise the amount of interwiring between the component panel and the controls, the tone control components are wired up pointto-point fashion on the tone control potentiometers. This wiring, together with details of the wiring to the other controls, is detailed in Fig. 3.

This is all quite straightforward, and due to the absence of a mains power supply and because the circuitry is screened by the case (which must be of all metal construction), it is not necessary for any of the wiring to have screened leads. However, all wiring should be reasonably short and direct.

Capacitors C13 and C113 require a chassis connection, and this is pro-

vided by a soldertag bolted to the top panel of the case just to the rear of VR3 and VR4.

PREAMPLIFIER PANEL

The magnetic pick-up preamplifier is assembled on a 0 lin matrix stripboard panel which measures 16 copper strips by 41 holes. Details of this component panel (board B) are provided in Fig. 2b. This is constructed in much the same way as the main component panel, and when completed under the top panel of the case to the rear of VR1, VR2, and S1. Wiring details are shown in Fig. 3.

IN USE

The completed amplifier requires no adjustment and is ready for immediate use. The amplifier is connected to the other items of equipment by way of twin screened leads terminated in the appropriate type(s) of plug. These input leads must be screened (the outer braiding connecting to the chassis of the amplifier) or there will almost certainly be a significant pick-up of mains hum and other signals in this wiring.

Many people tend to set the volume level rather high when they are using headphones, and it is advisable to avoid doing this as it simply results in increased distortion, and any imperfections on the programme source will become much more noticeable. Ц

EE CROSSWORD No 25 BY D. P. NEWTON

1

ACROSS

- 6 Street-like diode characteristic (3, 3)
- 7 Firmly closed door with a logic circuit (7)

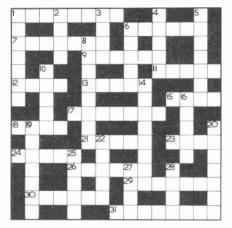
- **11** Currently a bog (5)
- 12 Sovereign conductor (4)
- 13 A circuit operating on the principle of least distance (5) To support a front for the e.m.f. (4) 15
- 17 Noise associated with random
- electrical discharge (7) Trace out an image electronically (4)
- Musical shorthand (5)
- 23 Diminutive designation of desk computer (4)

- 24 Solid follower of semiconducting circuits (5) 26
- Severe tension rising initially from each(6)
- 29 Circuit not resonating (7)
- 30 Oscillation exponentially dies away (6)

DOWN

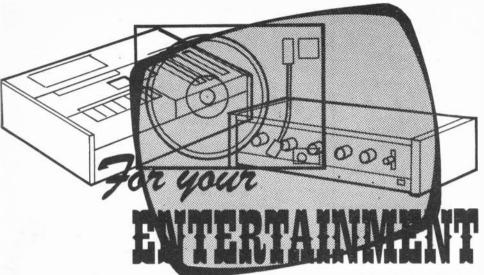
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- 2 Pitch a bit, perhaps caused by 28 down (4)
- Loud game? (Anag.) (4) 3
- Sounds like a result of interference 4 (5)
- 5 Microphone, full of energy (7)
- R.F. foe gives in worship (Anag.) (5)
- 8 Process of electron production (8)
- 10 Primary designator of amplification
- (5)14 Element (8)



- 16 10 down might result in this, once more (5)
- 19 Not positive about its connections (7)
- 22 Giant with a taste for humans (5)
- 25 Try a piece of prose (5)
- 27 0-0 (4)
- 28 Devices which rabbit about conversations (4)

Solution on page 184



A Clever Switch

There are all manner of anti car theft gadgets on the market, but many electronically minded owners simply put a hidden on/off switch in the ignition circuit. The engine then stubbornly refuses to start for anyone but its rightful owner.

Unfortunately thieves, (and for that matter the police, who drive offending cars off to, the car pound) are pretty skilled at finding secret switches. So here's a thought for car owning readers to ponder. Try putting not one, but two or even three, on/off switches in the ignition circuit. (Such switches should, of course, always be put in the low voltage supply lead to the coil *not* in the HT leads).

If you put more than one switch in the low voltage line and make one switch much easier to find than the others, you can create a confusing situation for wouldbe thieves. Try, for instance, leaving the most easy to find switch in the "on" state and a more difficult to find switch in the "off" state.

Anyone trying to start the car and realising that the ignition must be switched off will hunt for a hidden switch. They will then change its state. If the located switch was originally in the "on" state it will now be in the "off" state, so the ignition line will now be broken in even more positions than before.

If the unwelcome visitor to your car then hunts for, and finds, another switch the chances are that he will very soon become very muddled over which switch was in which position to begin with. With a bit If luck he will give up and try and start someone else's car. This is of course the object of the whole anti-theft exercise.

Off the Record

A few month's ago I wondered in this column whether any reader knew about the Tarbel computer system. The question arose because Isao Tomita, the Japanese electronic music composer, had written on the sleeve of his latest record, *Bermuda Triangle*, that the music contained a secret message.

This, he said, was coded in data that could be recovered by interfacing the hi-fi system with a computer "programmed

By ADRIAN HOPE

to the Tarbel system". But no one in the UK, even dedicated computer freaks, seems to know what the Tarbel system is.

Well partial relief is finally at hand. Tor-Arne Gisvold of the Norwegian Broadcasting Corporation (the Norwegian equivalent of the BBC) has now written to the monthly sound and music magazine *Sound International* (whose editorial staff were also curious and puzzled over Tarbel) with what seems to be the answer.

Tarbell (Tomita spelled it wrong) is, as we all suspected, a standard for recording digital data streams, not a computer programming system. But it differs from the most common Kansas City standard because it uses phase encoding at a frequency of 1496 bits per second.

Unfortunately no one in the UK seems to have any Tarbell interface gear so we still don't know what Tomita's coded message actually says. It can only now be an anti-climax. But he's certainly got some extra publicity for the record which was presumably one of the objects of the exercise anyway.

Alert and Well

One fast selling electronics item in the USA is the "air ioniser". For around \$100 you can buy a small cylinder that plugs into the mains and produces—or at least is claimed to produce—negatively charged air ions which make everyone in the vicinity feel alert and well.

But there are moves to curb the advertisement and sale of such gadgetry because it can constitute a health risk.

The resons for this are quite complex and we need to consider some basic electrochemistry to understand them.

Charged Molecules

lons are molecules of gas which carry a positive or a negative electric charge. In normal clean air there should be an equal number of negative and positive ions. But in fact, because negative ions tend to be repelled by the earth's surface, even normal air at ground level has a slight surplus of positive ions.

In polluted air the ion concentration drops and there is a greater surplus of positive ions over negative ions. This makes people tired, depressed and irritable. It isn't just polluted air that has an unwelcome imbalance of ions. When the Fohn wind in Germany and the Sharav in the Middle East blows almost everyone in the vicinity starts to feel "under the weather."

In some German cities hospitals cancel all but the most urgent operations when a local "ill wind" blows and causes a positive ion surplus.

What's more a surplus of negative ions does seem to have the opposite effect, making people feel bright, alert and generally pleased with life. Stand by a waterfall, where there tends to be a high concentration of negative ions, and you will feel positively glad to be alive.

A Possible Solution

The obvious solution to ion imbalance, then, is to produce an artificial boost to the negative ion concentration. This is what ion generators are intended to do. A considerable amount of research work on this subject has been carried out in Hungary, to stabilize the ion concentration for instance in a transport driver's cabin.

Hungarian medical workers found that in a large city the air in the driver's cabin of a bus can be massively overloaded with positive ions and clearly it is potentially highly dangerous to have a vehicle driver trapped in an atmosphere that makes him bad tempered. So they designed negative ion generators for use in road vehicles.

Ion Generator

On the face of things it seems an easy task to generate negative ions artificially. You simply push a high negative voltage into an electrode, and produce a corona discharge which negatively charges the air molecules. This is indeed the working principle of most ionisers which are currently sold for use in homes and offices or in cars.

A transformer system winds up the mains or battery supply voltage to levels which will produce a corona discharge from an ionising electrode. But this is where the problems begin. You don't just need a high voltage, you need a high charge density on the electrode.

For ionisation a minimum of around 4.5 kilovolts is needed at the tip of a needle electrode. But for safety you need around 10kV because as the electrode is used it will erode, becoming blunter and effectively larger and thus lowering the charge density for the same applied voltage.

Ozone

The real snag is that a low charge density on the electrode, will produce ozone rather than thoroughly ionized air. Ozone is really just partly ionised oxygen and it's a thoroughly unpleasant material. It's harmful to our throat and lungs, it kills plants, it destroys plastics and it may even cause cancer. The only saving grace is that ozone gives its presence away with the characteristic "electric" smell you get near sparking equipment.

It's the generation of ozone by inefficient ion generators which is worrying USA health authorities. So DIY constructors, or anyone owning a cheap ionizer, should be careful not to use anything that produces the characteristic smell of ozone. Responds to many types of gases. Will also detect accumulations of smoke.

★ Mains version for domestic use.

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S O FAR in this series we have looked at two properties of electronic components: resistance and capacitance. In fact, every circuit element which is **passive**, that is does not have any amplification associated with it, can be defined in terms of only three basic properties: two of them we have already discussed, the third being **inductance** which forms the first subject of this month's part.

When we have understood these three properties we can go on to look at circuits where a combination of the three types of components (resistors, capacitors and **inductors**) may be used to perform useful tasks.

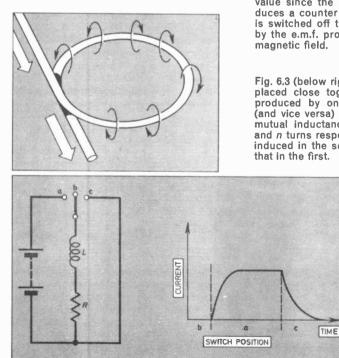
INDUCTANCE

As we saw in the first part of the series, when current flows in a wire it creates a magnetic field similar to that which is produced by a common bar magnet. If a piece of magnetised material is placed in this field then it will tend to be either attracted or repelled depending on the polarity of the two fields.

If we form the wire carrying the current into a loop thus making a coil with a single turn, the magnetic field will be as shown in Fig. 6.1. We can add more turns to the coil to increase the magnitude of the field, but notice that now each coil will now not only be in the field created by itself but also in the field created by nearby coils.

In 1831 Michael Faraday established a physical law which states that a moving magnet (or, more generally, a changing magnetic field) will induce an e.m.f. in a coil which cuts that field, and that the e.m.f. will be proportional to the rate of change of the magnetic field.

When current in a coil starts to build up from zero it creates an



increasing magnetic field around the coil. Now this field is changing and, as we discussed above, this will induce an e.m.f. in the same coil. In actual fact the e.m.f. which is produced is in such a direction as to oppose the e.m.f. which is creating the current in the coil.

When the current in the coil begins to decrease from some previously attained level, the magnetic field will decrease also. Again, the changing field creates an e.m.f. but this e.m.f. tends to maintain the current which is already flowing. Fig. 6.2 shows the rise and decay of current in a coil —or "inductor", as this type of component is usually called.

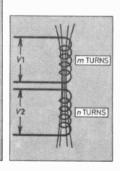
The property which causes these effects we call **inductance** and the units in which it is measured we call **henries** (H). To be a bit more precise, we say that a circuit has an inductance (or, more strictly, a **self-inductance**) of one henry if the e.m.f. induced in it is one volt when the current is changing at a rate of one ampere per second.

Henries are sub-divided into millihenries (mH) and microhenries (μ H) for practical use.

Fig. 6.1 (left). When a wire carrying current is formed into a single turn a magnetic field is produced as shown here.

Fig. 6.2 (below left). When a voltage is applied across an inductor it will take a time for the current to attain its final value since the increase of current produces a counter e.m.f. When the voltage is switched off the current is maintained by the e.m.f. produced by the collapsing magnetic field.

Fig. 6.3 (below right). When two coils are placed close together so that the field produced by one envelops the second (and vice versa) we have what is termed mutual inductance. If the coils have m and n turns respectively, then the voltage induced in the second will be n/m times that in the first.



MUTUAL INDUCTANCE

If two coils are placed next to each other the changing magnetic field from one of them will produce an e.m.f. in the other.

If the two coils were identical and it could be arranged that the field produced by the first totally enveloped the second coil, then the e.m.f. produced in the second coil would be exactly the same as the e.m.f. producing the current in the first coil (see Fig. 6.3).

We call the property which induces a voltage in one coil from another mutual inductance. It has the same units as self-inductance, henries. We define mutual inductance by stating that two circuits have a mutual inductance of one henry if the e.m.f. induced in one of them is one volt when the current in the other is changing at one ampere per second.

TRANSFORMERS

If the second coil is totally enveloped by the field of the first but this time has more turns than the first, then the voltage induced in it will be exactly proportional to the number of turns in the second divided by the number of turns in the first. We call a device which takes advantage of this property a **transformer**.

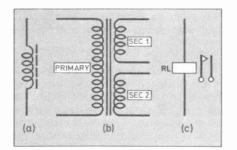


Fig. 6.4. The symbols for three commonly encountered inductors. (a) shows an inductor, (b) transformer and (c) relay with a single pair of normally open contacts. In the case of (a) and (b) the lines next to the coil indicate the type of core.

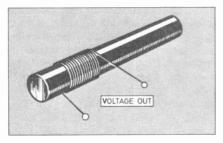


Fig. 6.5 A transistor radio aerial consists of a ferrite rod around which is wound a coil (or several coils). The radio waves induce small e.m.f.s in the coils.

Notice that a transformer requires there to be a *changing* current in one of its coils in order to induce an e.m.f. in the other coil (or coils, for there may be many). We can thus only use the transformer in a.c. circuits.

Of course, the mains electricity supplied to our houses is alternating and we can therefore use a transformer to "step down" the high mains voltage to a more useful low voltage simply by having a large number of turns on the coil connected to the mains and a small number on the side from which we want low voltage.

It is general practice to refer to the winding into which power is fed as the "primary" winding and the windings out of which power is taken as the "secondary" windings.

In transformers we often find that we have secondaries connected together to form what is termed a centre-tapped winding. The transformer used on the Tutor Deck has in fact two separate secondaries but the start of the first coil has been connected to the finish of the second to form a centre-tapped winding.

It is important to connect the coils in the correct sense otherwise the e.m.f. induced in one of the coils will oppose that induced in the other. Thus if the two "starts" were joined we would get no e.m.f. across the two "finishes".

The thickness of the wire with which the transformer primary and secondary are wound depends on the current which will be flowing through the coils. In a perfect transformer the power put into the primary would exactly equal power drawn from secondaries.

Thus if the transformer was a step-down type (fewer secondary windings than primary) then to maintain the power in-out balance the secondary voltage times the secondary current must equal the primary voltage times the primary current. $(P = V \times I)$.

TURNS RATIO

If the turns ratio is n:1 (where n is greater than one) then the primary voltage must be n times the secondary. This implies that the secondary current must be n times the primary.

In general, for a step-down transformer as used in low-voltage power supplies the secondary will be wound with thicker wire than the primary in order to carry the higher current.

Transformers for power supplies are usually specified in terms of their primary voltage and their secondary voltage(s) and current(s). The symbol for a transformer is in Fig. 6.4 (b).

INDUCTORS

Inductors as circuit components are used less and less in modern times. They are avoided because they tend to be bulky for anything other than very small values of inductance; they tend to be sensitive to external magnetic fields and also to create unwanted magnetic fields.

One area where coils are still used extensively is radio circuits, in particular in the aerial circuits. Most transistor radios use a **ferrite rod aerial** which is wound with a number of coils each designed to be particularly sensitive to a given range of radio frequencies.

The ferrite core concentrates the electromagnetic energy of the radio waves so that there is a relatively large changing magnetic field affecting the coil. The radio waves induce tiny e.m.f.s in the coils which are subsequently amplified and decoded into the original audio signal. See Fig. 6.5.

In a radio receiver there are usually, in addition, three or more transformers of a different kind. These are enclosed in metal cans made from a special material which prevents external fields from affecting their operation.

SOLENOIDS

The magnetic field produced by the current flowing in a coil is put to use in a number of **electromechanical** devices. **Solenoids** are simply coils wound round a core which is free to move. When the current is on, the core is pulled into (or pushed from) the coil and this movement can be used to actuate some mechanical device such as a lock.

When the moveable part is used to close or open an electrical switch we have what is termed a relay. The symbol for a relay is shown in Fig. 6.4(c).

When we are driving devices with inductance with circuits which are sensitive to high voltages we must be aware of the

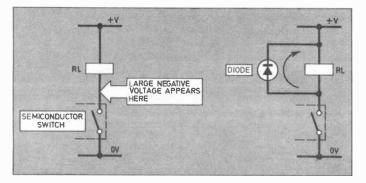


Fig. 6.7 (right). Diagram illustrating the terms period, frequency, amplitude and phase.

Fig. 6.6. If a voltage sensitive switch (such as a transistor) is used to switch on and off a relay then some protection must be provided against the back e.m.f. produced by the collapsing magnetic field. This is usually in the form of a diode which provides a path for the current produced by the e.m.f. in the coil.

high e.m.f. produced by the collapsing magnetic field produced when we switch off the current.

For instance the diagram in Fig. 6.6 shows a switch connected to a relay. After the switch has been on for a long time a constant current will be established in the coil. When the switch is turned off the magnetic field will collapse producing an e.m.f. which tends to maintain the current in the coil.

The voltage at the top end of the coil is prevented from altering since it is connected to the power supply rail, but the bottom end will go to a very high negative voltage. If the switch is some sort of semiconductor device then it will inevitably be destroyed.

A number of remedies are possible, the simplest of which is to connect a diode across the coil with the cathode to the positive supply (Fig. 6.6(b)). The diode now provides a path for the current when the field collapses.

Because the voltage across the coil is now clamped to the forward voltage drop of the diode the energy stored in the inductance takes longer to be dissipated. This is the penalty we must pay, but in most cases the slowing up is irrelevant.

Like capacitors, inductors are stores of energy. But whereas capacitors store the energy in the form of electric fields, inductors store energy in magnetic fields.

A.C. CIRCUITS

Now that we have looked at the three basic elements which go to make up every real circuit component, we must see how the three elements interact in real circuits.

When we are analysing circuits with repetitive waveforms we

make use of the fact that any real repetitive function can be thought of as composed of a number of pure sinewaves of varying amplitude and phase added together.

The terms **frequency**, **amplitude** and **phase** need to be defined and to do this we refer to Fig. 6.7.

The figure shows a sinewave starting at time zero. After a time t the waveform begins again and repeats exactly the shape of the first "cycle". The time between two identical points on a waveform we call the **period**.

The number of cycles in one second we call the "frequency". The frequency is the reciprocal of the period.

f = 1/t

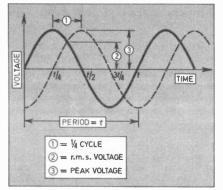
Frequency is measured in hertz (Hz) where 1 Hz = 1 cycle per second.

AMPLITUDE

The amplitude of the waveform can be defined in a number of ways. We can simply define the peak amplitude as the voltage at the top of the waveform. The more usual way of specifying the amplitude is to quote the **root mean square** (or **r.m.s.**) value. This is defined as the value of the d.c. voltage which would produce the same heating effect as the a.c. waveform.

For a sinewave there is a simple relationship between the peak value and the r.m.s. value (r.m.s. amplitude = $0.707 \times$ peak amplitude) but for anything other than this the relationship is more complex.

The waveform shown dotted in the diagram has exactly the same amplitude and frequency as the solid waveform but it is displaced in time. As we move from left to



right along the time axis we move on in time hence the dotted waveform is "lagging" the solid one.

Each complete cycle can be divided into 360 degrees just like one complete revolution around a circle. We can specify by how much the dotted waveform is lagging the solid by using degrees as our unit. In the case of the diagram, the dotted waveform is lagging the solid waveform by a quarter of a cycle or 90 degrees.

Using all these terms we can now look at how inductors, capacitors and resistors affect a.c. waveforms which are applied to them.

CAPACITOR PLUS RESISTOR

The diagram in Fig. 6.8(a)shows a sinusoidal waveform applied to a simple network consisting of a resistor and capacitor; we wish to discover how the output waveform differs from the input.

When we were discussing capacitors we noted that the current through the capacitor was equal to the capacitance times the rate of change of voltage across that capacitor. Now the total voltage across the capacitor plus the voltage across the resistor must at any particular instant equal the input voltage.

The voltage across the resistor and that across the capacitor are shown in Fig. 6.8(b) and many interesting points can be noticed.

First, the total voltage across the two components (the driving voltage) has a phase which is between that of the capacitor and resistor. Also, the maximum voltage is not the sum of the two separate peak voltages because these do not occur at the same time. However, it is larger than either of the two components.

Though a solution to the problem has been presented, the way in which it was derived has not. It is clear that deriving the voltages in any circuit containing capacitance (and indeed inductance) is not possible using simple arithmetic as was possible with circuits containing only resistance.

In fact when this subject is studied in more detail it will be found that there are two main methods of circuit analysis: one is a method using geometry (called the vector method) and the other using a special type of mathematical operator (called the "j-operator"). Neither of these methods will be described here, only the results obtained will be presented.

INDUCTANCE PLUS RESISTANCE

In a circuit containing resistance and inductance we again have a complex situation as the voltage across an inductor is proportional to the rate of change of current through that inductor. We can imagine the capacitor as being a store of voltage and the inductor as being a store of current.

If we apply an a.c. voltage to the inductance-resistance network then the voltages and currents which are produced are shown in Fig. 6.9.

Notice that the current now lags the phase of the voltage.

PART 6 QUESTIONS

6.1. What is the inductance of a coil which has 6mV across it when the current is changing at 3mA per second:

a) 0.5mH

b) 2H c) 2μH

d) 5H

u) 5m

6.2. A transformer has 250 primary turns and 25 secondary turns. What will be the secondary voltage when 250V are applied to the primary:

a) 100V b) 10V

c) 25V

d) 2.5V

6.3. For the transformer described in 6.2, a current of 2A is drawn from the secondary.

THE TUNED CIRCUIT

When inductance and capacitance are combined together in a circuit we have what is termed a **tuned circuit**. We are all familiar with the process of "tuning in" a radio station: we simply turn a control until the station that we want to hear is at its clearest; either side of this setting the reception is poorer.

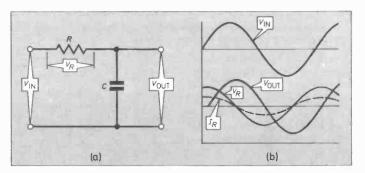


Fig. 6.8. When the capacitor-resistor network at (a) is fed with the alternating voltage, the voltages and currents are as at (b) the current leading the voltage by 90 degrees.

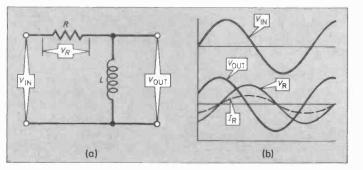


Fig. 6.9. When the inductor-resistor network at (a) is driven with an alternating voltage, the voltages and currents produced are as at (b) the current lagging the voltage by 90 degrees.

What is the primary current:

a) 1A b) 0·1A

- c) 0.2A
- d) 20A

6.4. What is the resonant frequency of a 200pF capacitor in parallel with a 10μ H inductor:

- a) 20MHz
- b) 3.56MHz
- c) 35.6MHz
- d) 200kHz

6.5 When a 200pF tuning capacitor is set to 63.3pF a radio is tuned to 200kHz. What is the parallel inductance:

- a) 10mH
- b) 63.3mH
- c) 200mH
- d) 10µH

PART 5 ANSWERS

5.1. a) 5.2. a) 5.3. b) 5.4. a) 5.5. a)

The tuned circuit has a response which varies in a similar way as the frequency of the driving signal is altered.

There are two ways of connecting the capacitor and inductor: in series and in parallel.

We said earlier that the capacitor could be regarded as a store of voltage and the inductor as a store of current. When the two elements are connected there is a transfer of energy between the two. At a particular frequency, called the **resonant frequency**, the two storage elements act together in a special way.

When the capacitor and inductor are connected in parallel we know that the voltage across the two elements must be the same. Now the current through the capacitor leads the voltage while the current through the inductor lags the voltage. At the resonant frequency these two currents cancel each other out and the capacitorinductor combination effectively acts as a barrier to currents of the frequency to which it is tuned.

Of course, this assumes perfect components: inductors without resistance or capacitance, and capacitors without resistance or inductance. Real components will have a little of each, although the most significant "unwanted" component is the resistance of the coil used to make the inductor. The effect of this is that the capacitor and inductor are not exactly out of

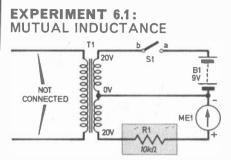


Fig. 6.12. The circuit of Experiment 6.1 with the layout of the components shown at (b).

The circuit diagram for the experiment is shown in Fig. 6.12a and the layout on the Tutor Deck in Fig.6.12b. Note that the transformer is not plugged into the mains—we are only using the two secondary windings to show how the change in current in one produces an e.m.f. in the other.

When the switch S1 is open no current can flow in the first winding but as it is closed current builds up. The time it takes to reach the final value and the magnitude will depend on the resistance and inductance of the coil, but these actual values are not important.

EXPERIMENT 6.2: SIMPLE POWER SUPPLY

Components needed: 1N4148 diodes (2 off), 5·1V Zener diode, 470 μ F 25V capacitor, 1k Ω $\frac{1}{2}$ W resistor, 680 Ω resistors (2 off).

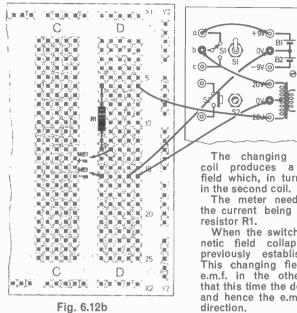
Using the principles so far covered in this series one can combine components to form a simple power supply which can be used to drive any circuits requiring 5V at a few milliamps.

The circuit for the power supply and the layout on the Tutor Deck are shown in Fig.6.13a and Fig.6.13b.

The diodes D1 and D2 are used to form a full-wave rectifier. This produces pulsating d.c. and to smooth it we use a high value electrolytic capacitor C1. The capacitor charges up to the peak voltage produced by the diode bridge. As the voltage from the rectifier starts to fall the voltage stored on the capacitor reverse biases the diodes so that the only path for the current is into the circuit to the right of the capacitor.

The capacitor is followed by a 340 ohm resistor in series with a $5 \cdot 1V$ Zener diode D3. This resistor is actually two 680 ohm resistors connected in parallel (R2, R3). The value is chosen so as to pass sufficient current into the diode to keep it on its constant voltage operating characteristic yet allow enough current to pass into the load, this being connected across the Zener diode.

With values shown the circuit can deliver approximately 10mA at $5 \cdot 1V$. If a load which demands more current than this is connected across the output, then the voltage across the resistor will increase and the Zener diode will no longer receive enough current to keep it in its constant voltage, mode, hence the output voltage will drop.



The changing current in the first coil produces a changing magnetic field which, in turn, produces an e.m.f. in the second coil.

The meter needle will be deflected, the current being limited by the series resistor R1.

When the switch is opened the magnetic field collapses as the current previously established falls to zero. This changing field again induces an e.m.f. in the other winding but note that this time the deflection of the meter, and hence the e.m.f., is in the opposite direction.

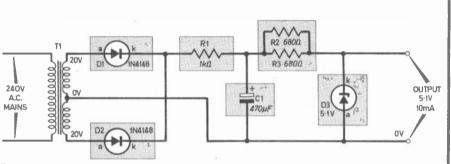
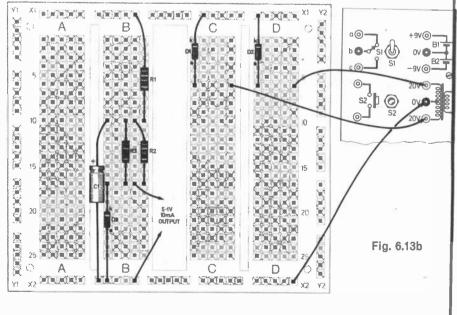


Fig. 6.13. Circuit of the simple power supply which is the subject of Experiment 6.2. The layout of the components on the Tutor Deck is shown at (b).



phase, and so a small current flows even at resonance.

The plot of current against voltage for varying frequencies in a parallel tuned circuit is shown in Fig. 6.10. The actual relationship between the values of the capacitance and inductance and the resonant frequency are

$$fr = \frac{1}{2 \times \pi \times \sqrt{L \times C}}$$

In a series tuned circuit as shown in Fig. 6.11 we have the situation that the current through the capacitor and the inductor are the same. At the resonant frequency two equal and opposite voltages are produced across the capacitor and inductor and these tend to cancel so maximum current flows through the network. The effect is that the total network appears like a short circuit.

Again the above description assumes perfect components; in practice even at resonance there will be some resistance in the circuit.

The variation of the current flow through the series network is shown in Fig. 6.11(b). The relationship between the resonant frequency and the component values is exactly the same as the parallel circuit.

USES OF TUNED CIRCUITS

Tuned circuits find use most often in radio circuits where their characteristics make them ideal for selecting the particular radio

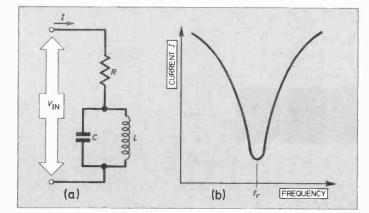


Fig. 6.10. A parallel resonant circuit is shown at (a) and the relationship between frequency and current at (b).

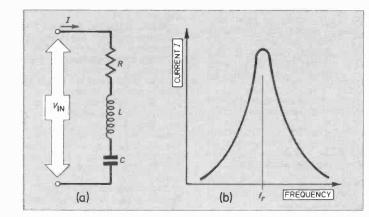


Fig. 6.11. A series resonant circuit is shown at (a) and the relationship between frequency and current at (b).

frequency in which we are interested. One of the components (usually the capacitor, but in higher frequency systems the inductor) is made variable so that the resonant frequency of the network can be varied at will. The other use of tuned circuits is in oscillators and we will see how these operate in a later part of this series.

Next month we will meet that very important semiconductor device, the transistor.

DAD'S INSTALLED THE LOFT ALERT SYSTEM I KNOW. ALL WE NEED NOW IS A SYSTEM TO WARN US... WHEN HE'S GOING TO PUT HIS FOOT THROUGH THE CEILING!

TOTAL



GINNERS

TABLE TOP OPERATIONS

One of the great attractions of elec-tronics construction is that one does not need special or elaborate workshop facilities in order to get started in this hobby. A table or any flat surface is all that is needed, essentially, for assembly and wiring-up electronic circuits.

There are of course some objections to working directly on the table top. Possible damage to the surface is naturally one; another is the incon-venience of having to gather up all one's bits and pieces when the table is needed for other purposes. A simple solution is to obtain a sheet of hard-beard, say 2ft by 3ft and fix wooden battens around three edges to retain the tools and components.

The hardboard work top can be without any disturbance to any partly completed job in hand. The board may also be found suitable for working in an armchair. There's luxury for you!

One possible hazard to be guarded against is damage caused by a hot soldering iron.

Always have a suitable metal receptacle at hand and place the iron in this when not in use. Some soldering irons have their own "stand", but it is still advisable to have a metal tray or suchlike over which the iron should be held when wiping or tinning the bit.

A small piece of expanded foam, a thin sponge pad, or a duster folded several times, makes a suitable mat for the circuit board during component assembly and soldering operations.

The purpose of a mat is twofold. It provides a suitable grip for the board, so preventing it moving at a critical moment during soldering operations. It cushions the components from the working surface—thereby preventing damage to the components themselves while also protecting the working surface from scratches.

The metal tray has a further use. When component leads have to be cut (cropped) perform this operation over the tray. At the end of a building session, it is a simple operation to empty the tray into the dust bin. Adopt this procedure and there should never be complaints of solder globules or tiny bits of wire embedded in the

carpet. When building a project some light engineering or mechanical work is sometimes involved. For example, cutting and bending a piece of alu-minium, or drilling holes in metal or s.r.b.p. This kind of work should certainly not be attempted on the table top. Here one should of course resort to the normal procedures and carry out this kind of work in the shed, garage or room specifically allocated for workshop activities. In any event, a mechanics vice is likely to be essen-tial for this "more physical" kind of work.

Thus the major part of project building can be carried out in agreeable surroundings, without damage to property or hindrance to other occu-pants of the room.

As one becomes more experienced and ambitious, the need for a special workbench or workshop area will probably arise. For one thing, there will be a few test instruments and a power supply unit to make provision for, in due course, as well as an in-creasing stock of electronic com-ponents. But until this stage is reached, the newcomer can enjoy him-self no end using the minimum amount of equipment and making use of the table top.

BASIC TOOLS -

Listed here are minimum requirements in the way of tools and equipment for electronic construction.

Screwdriver 2mm (Lin) blade Screwdriver 4mm (3/16in) blade Miniature side-cutters Miniature insulated long-nose pliers Side-action wire strippers Utility knife

Miniature soldering iron, 15-25 watt 2 bits: 2.3 and 4mm (approx.) Cored Solder (60% tin, 40% lead) 22s.w.g. Small Multimeter:

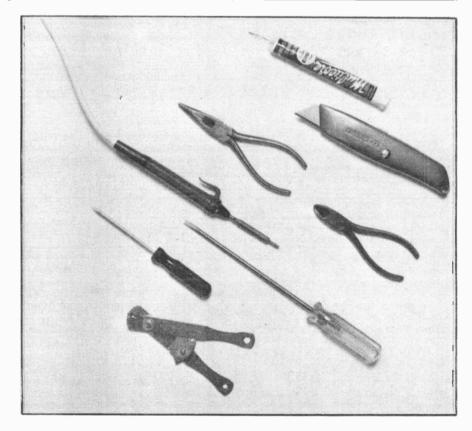
Ranges (minimum specification)*

D.C. 10, 50, 250, 1,000 volts A.C. 10, 50, 250, 1,000 volts D.C. 1m A, 100m A

Resistance 0-150 kilohms

*for example Type FL60Q (Maplin Elec-tronic Supplies Limited)

Next month a suggested list of circuit components for the newcomer to purchase will be given in Square One.





Less than five years ago, the thought of owning one's own digital computer would for most people have seemed pretty remote. Actually building such an intricate and powerful machine would hardly have been contemplated by the private individual due to the complexity and sheer volume of construction involved with the components available at the time—and the cost.

With the advent of the microprocessor all this has been changed. Tens of thousands of digital computers are privately owned in the UK alone and sales are accelerating.

The majority of these have been purchased ready built and tested but a large number have been built from a microcomputer kit. One such kit is for the TRITON, designed and marketed by a British firm called Transam. Since its launch in November 1978 more than 2,000 kits have been purchased.

It is not necessary to understand the internal workings of a computer to be able to use it to maximum advantage. But for those who do want this knowledge, we believe one of the best ways is to actually put a system together in conjunction with the theory of the system as provided in this package. It often helps to understand more fully the working of devices if you know what they look like and how they are physically interconnected to others. Besides this, great satisfaction is obtained from "doing it yourself".

SINGLE BOARD

Nearly all the components in the Triton are mounted on a ready drilled (double-sided fibre-glass tinned) printed circuit board with plated through holes, as can be seen in the photograph. The board is guaranteed free from design errors.

By far the most important requirement for anybody thinking of building this system, as any other electronic project, is the ability to solder. There is no danger however of i.c.s being damaged by the soldering iron since all i.c.s are mounted in sockets. The fitting of these sockets forms a major part of the construction on this board and requires no special tools, only a good miniature soldering iron. Construction appears to be straightforward and explained in detail in the 116 page manual.

One criticism we have regarding the presentation of the constructional information, is the lack of constructional and interwiring diagrams that readers of EE can take for granted. Only one drawing is provided which is rather cramped. It shows the total component layout of the p.c.b. overlayed on the printed circuit pattern. With such a project it would be desirable to have more interwiring diagrams. There are none for the keyboard, power supply or control switches. A stage-by-stage assembly diagram would be most helpful.

KIT OF PARTS

The Triton kit comes complete with power supply and a full 56 station ASCII keyboard to be mounted in a pre-drilled and formed aluminium case (not shown) which has a large flat horizontal top panel on which the TV can stand (VDU). The photograph shows the alternative vacuum formed rugged plastic case available.

For those on a limited budget, the basic kit has been broken down into 16 individual packs available separately enabling the system to be purchased over a period as funds permit. Also the constructor has the option of using components he may already have in stock. Needless to say, the total cost of the individual packs exceeds the total kit price.

The Triton uses the well tried and tested Intel 8080 microprocessor. Total on-board memory is 8K of which 3K is user RAM. The memory can be easily increased via an external motherboard system up to the full 64K, in blocks of 8K pluggable RAM cards and/or 8K EPROM cards. Both are available in kit form.

The system has been designed to plug directly into the aerial socket on a standard UK TV, channel 36. A cassette interface is included allowing programs/data to be rapidly loaded in and out of the computer using a standard cassette tape recorder.

FIRMWARE

All of the firmware is resident in UV erasable EPROMS. There are six Basic firmware options to choose from to suit most needs: L4.1 (1K monitor 2K Tiny Basic having seven monitor functions); L5.1 (1.5K monitor 2.5K Tiny Basic with 14 monitor functions); L6.1 (2K monitor 7K Basic with 25 monitor functions) this is a scientific package with full floating point arithmetic; L7.1 (8K Extended Basic). Also available are L4.2 and L5.2, higher clock frequency (18 MHz) versions of L4.1 and L5.1. In addition a Pascal firmware pack L8.1 has recently become available.

The basic Triton L4.2 kit (£286 + VAT) is an excellent introduction to the fascinating world of computers and can be easily adapted to suit requirements and expanded up to a 64K system with full Basic or Pascal by the addition of the motherboard and plug in RAM/ROM cards.

Firmware L9.1 has recently become available to allow the Triton to be interfaced with a floppy disc system and there is more to follow, as 50 per cent of Transam's investment is dedicated to developing new soft/ firmware.

An after sales service exists for constructors who have experienced difficulties in getting the completed system operational. Also for those people not sufficiently skilled in the art of soldering or construction, the microcomputer may be purchased ready assembled at an extra cost of about £60.

If you require further information we suggest that you contact Transam Components Ltd; Dept. EE, 12 Chapel Street, London NW1.



OOK REVIE

SYMBOLS AND ABBREVIATIONS FOR ELECTRICAL AND ELECTRONIC ENGINEERING **IEE Editorial Panel** Editor Price £0.75 210×148 mm 16 pages Size Publisher Institution of Electrical Engineers ISBN 0 85296 197 7

THIS IS A fully revised edition of a similarly entitled

booklet previously published in 1968 and 1971. It contains the IEE recommended usage for students and staff in educational establishments. Also it is intended as a handy reference for authors of papers and books on technical subjects, and for draughtsmen and designers in industry. It takes into account the latest relevant British, Military and International Standards.

Rules governing the formation of abbreviations for words and phrases are explained with examples. The International System (SI), derived units, mathematical symbols, subscripts and conversion factors are included. Two pages are devoted in circuit symbols.

» % * » ** * 232°98 . 38982 10 m SINGLE I.C. PROJECTS Author R. A. Penfold Price £1.50 180×108 mm 127 pages Size Publisher Bernard Babani ISBN 0 9001 6285 6 × > 1. w . M

IKE the curate's egg, one is tempted to say "good in parts". With the advent of specialised i.c.s construction of even quite complex devices has been made that much simpler and the 20 projects presented here reflect this trend.

As the title suggests, each circuit incorporates just one i.c. and provided a modicum of care is exercised, even a relatively inexperienced constructor should be able to produce a range of useful gadgets.

That is he would be able to if he had the patience to sift through the text, find the right diagram for the right design, work out which component list is which and so on.

The layout and organisation of the text and diagrams is somewhat haphazard to say the least and even the contents list does not reveal all the projects that are in the book.

Regular readers of EE will no doubt be familiar with the author's name and it is a great pity that the presentation does not do his designs justice. S.E.D.

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THE first two-thirds of this book merit its title. The final 100 pages are shared between a miscellany of electronic products: watches, calculators, home protection and security devices, watches, carchators, nome protection and security devices, amateur and CB radio hi-fi, and in-car entertainment. These brief surveys, the work of individual contributors, have been added, it must be supposed, to a predominately TV or video survey in an attempt to justify the sub-title "The Future of Consumer Electronics". But it falls short here, by at least one glaring omission-the microwave oven.

As a review of technical progress to date and a forecast of future developments in television and video in all its various forms this book is certainly a timely and useful work. Video recording, for example, is going to make its impact upon the domestic scene and now is the time for an appraisal of techniques involved. The chapters on video cassette recorders (Angus Robertson) and video discs (Adrian Hope) will help clear many misconceptions. Teletext and Viewdata are described and the future of these information corriging is predicted Television these information services is predicted. Television games get good coverage, but home computers receive but cursory treatment in mere nine pages.



Thanks for the Memory

I found Mr. Young's recollections of early radio, see Counter Intelligence January issue, very interesting but slightly erroneous.

Children often misjudge measurements which is evidensed when they return to childhood houses in later years and find them much smaller.

Grid bias batteries were certainly not a yard long about 9 inches I think. Also H.T. batteries could hardly have weighed half a hundredweight. I was no Samson and had to bring them home very fre-quently-they did not last long.

Accumulators were not that big either. My brother made a living re-charging them for radio owners and they were quite portable.

I certainly remember Leonard Henry and many others. There was the "Roos-ters Concert Party" which I found superb. They must have been the pioneers of sound effects, and reproduced marching and other forms of transport to perfection. They started in the first world war and my elders did not find them as good as I did—probably because of war-time memories. After a while they disappeared and I have not since found a concert party to equal them.

I won't bore you any further with my reminiscences but thank you for recalling such a happy aspect of my youth.

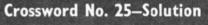
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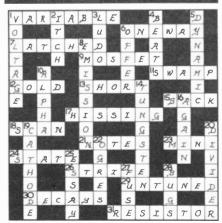
Peg-A-Hole

When constructing your project Peg-AHole game, I used the thyristor THY1A400. When using this I found that R1, a $2 \cdot 2k\Omega$ resistor was too large because it did not allow enough current to the gate terminal.

I found a 220 Ω the most suitable replacement.

> Paul Snape (Age 11) East Gomeldon







T CAN be useful to know whether someone has called when one is out of the house. The simple device described here will provide such an indication. It has other uses as well, and could be used as a general event recorder.

CIRCUIT

The circuit (see Fig. 1) is built around a cmos i.c., type CD4013. This is a dual flip-flop, only one half of which is used in this project.

Operation is as follows: A positive clock pulse applied to pin 3 causes the output at pin 2 to go from low to high and turn on the light emitting diode D1. The output stays high when further pulses are applied, until a positive pulse to pin 6 via S1 resets the output to low.

Components R1, R2 and C1 are included to give protection to the i.c. from the high voltages which can be generated from coils in the door bell.

Being a cmos chip the CD4013 only consumes about a micro amp when the l.e.d. in the off state, and about 7mA in the on state.

CONSTRUCTION

The prototype was built in a Vero box measuring about 45 x 100 x 25mm. The components, with the exception of S1 and S2, are all mounted on a piece of 0.1 matrix stripboard measuring 1.75 inches x 1.25 inches approximately (12 strips by 16 holes). Component arrangement and breaks in copper strips are shown in the diagram Fig. 2.

The lid of the box is drilled such that the l.e.d. protrudes through it when the lid is placed in position. S1 and S2 are mounted on either side of the battery. The layout is not at all critical and can be altered to suit the constructor.

The input leads from the bell are brought in via a hole drilled in the back of the case. The other end of these leads should be attached to the bell terminals.

The component board is secured to the box with one 6BA screw using a brass tapped-bush incorporated in the box.

When soldering the i.c. be sure to use a well earthed soldering iron, and never solder or insert the i.c. into the circuit with the power on.

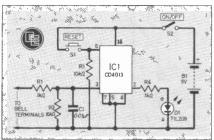


Fig. 1. Circuit of the Doorbell Register.

The 9V battery sits securely in the lid, being a tight fit between the pair of tapped bushes.

When the instrument is first switched on the l.e.d. may light; press the S1 to reset it. If the l.e.d. does not light at all, it is probably wired in with its terminals reversed.

COMPONENTS

S1 is a miniature push-to-make type switch. S2 is a miniature on/off switch. Any small l.e.d. can be used, but be careful to get the right polarities when assembling.

The device will work on any voltage from 3V-15V. A PP3 is a convenient size of battery. If a lower voltage is used, for example 3 volts, reduce R1 to 270 ohms.

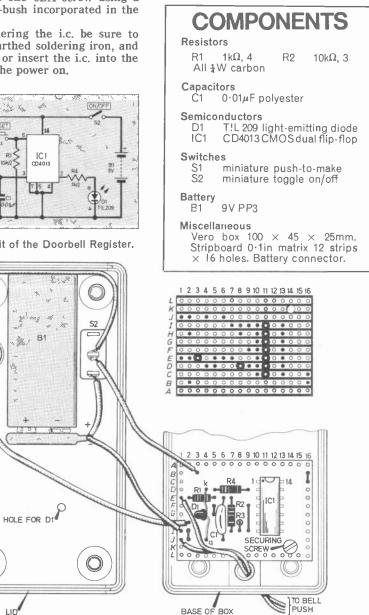


Fig. 2. Interior view of box showing assembly of parts and wiring between lid and base of box. Top right: underside view of stripboard.

RANGE CURRENT LIMITER

THIS type of "electronic fuse" protects d.c. power supplies against short circuits or excess loads, as well as limiting the current which can flow in equipment operated from the supply. There are five ranges, selected by a switch. These allow limiting at 50mA, 100mA, 250mA, 500mA and 1 ampere.

CIRCUIT DESCRIPTION

The complete circuit diagram of the current limiter is shown in Fig. 1. Transistor TR1 is a d.c. amplifier, TR2 the series control transistor, and TR3 the limiting circuit transistor. Switch S1 selects the wanted range, and the limiting action depends on the values of the resistors R2 to R9. By employing resistors in parallel, standard values of ${}^{1}_{4}W$ resistors can be used

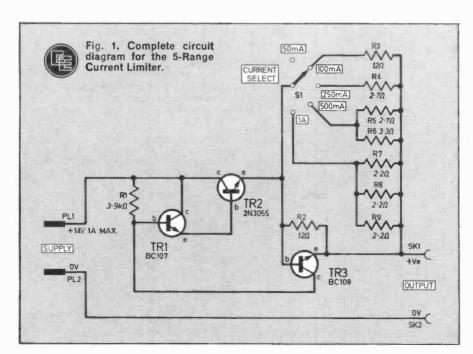
throughout, which is an advantage when obtaining the components.

Actual values for the five ranges are $(12 \text{ ohm } (R2) \text{ for } 50\text{mA}, 6 \text{ ohm} (R2 \text{ and } R3 \text{ in parallel}) \text{ for } 100\text{mA}, 2 \cdot 2 \text{ ohm } (R2 \text{ and } R4 \text{ in parallel}) \text{ for } 250\text{mA}, 1 \cdot 32 \text{ ohm } (R2, R5 \text{ and } R6 \text{ in parallel}) \text{ for } 500\text{mA}, \text{ and } 0 \cdot 69 \text{ ohm} (R2, R7, R8 \text{ and } R9 \text{ in parallel}) \text{ for } 1A.$

CURRENT LIMITING

With current under the set limiting value, TR1 base is supplied by R1 and TR2 base from TR1 emitter. Transistor TR3 is not operating since the emitter-base potential depends on the particular resistance value present between these two points, for example 12 ohm for 50mA. Current is thus available through TR2 for the equipment. GOMPONENTS approximate GOSt £3.50

Should current increase due to a heavy load or short circuit, the voltage drop across R2 rises to 0.6V for 50mA, so TR3 switches on. The excess load or short circuit across the output terminals means that TR3 emitter is



COMPONENTS
$\begin{array}{c} \text{Resistors} \\ \text{R1} 3 \cdot 9 \text{k} \Omega \\ \text{R2} 12 \Omega \\ \text{R3} 12 \Omega \\ \text{R4} 2 \cdot 7 \Omega \\ \text{R5} 2 \cdot 7 \Omega \\ \text{R6} 3 \cdot 3 \Omega \end{array} \qquad \qquad \begin{array}{c} \text{See} \\ \text{Shopp} \\ \text{Talk} \end{array}$
R7 2·2Ω page 167 R8 2·2Ω 2 R9 2·2Ω 2 All ½ watt carbon ± 5% 5%
Semiconductors TR1 BC107 npn silicon TR2 2N3055 npn silicon TR3 BC108 npn silicon
Miscellaneous S1 2-pole, 6-way switch(see fig.2.) SK1, SK2 4mm sockets (one red, one black) PL1, PL2 4mm plugs (one red,
one black) Metal box 165 × 65 × 40mm; five- way tag strip; mica washer, insu- lating bushes and mounting nuts and bolts for TR2; control knob; solder tag; two lengths of p.v.c. covered 16/0.2mm connecting wire (one red, one black).

negative, so conduction in TR3 moves TR1 base negative, and TR1 emitter follows this, shutting off TR2.

For heavier currents, R3 or other resistors will be in parallel with R2, as described. The current level at which TR3 operates is thus increased.



ASSEMBLY

A metal box $165 \times 65 \times 40$ mm is used to house the unit. Transistor TR2 is mounted on the side of the box itself using insulating bushes and mica washers and the remainder of the components are wired between a fiveway tag strip and SI as shown in Fig. 2.

The input to the unit from an external power supply is taken via red and black flexible leads terminated by plugs PL1 and PL2, the output is taken from sockets SK1 and SK2.

Resistor R2 is in circuit for all ranges, so no connection is made to the 50mA position tag on S1. The unit could simply be placed in the positive supply line, but it was found more convenient generally to have negative run to the second socket.

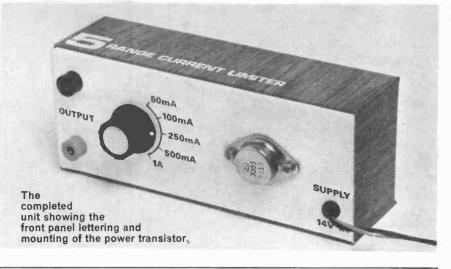
TESTING THE LIMITER

After assembly, the unit can be checked by placing a meter in series with the supply. Set the instrument to the 1A or larger current range, or to an appropriate lower current range for the lower current positions of S1. With a deliberate overload or short circuit at the sockets of the unit, current will be limited to about the values shown for the various switch positions.

USING THE LIMITER

The unit is intended for power supplies of up to 24V. In the worst operating conditions, that is a complete short circuit of output on the 1A range with 24 volts, dissipation in TR2 is about 24 watts. This is not immediately too important because the panel and box are a heat sink for TR2, but the short or overload should of course be removed.

At lower currents and voltages, the heating of TR2 will be proportionally less. Switch S1 is set to that current range which comes most nearly above the normal current demand of the equipment, model, or other apparatus being powered. \square



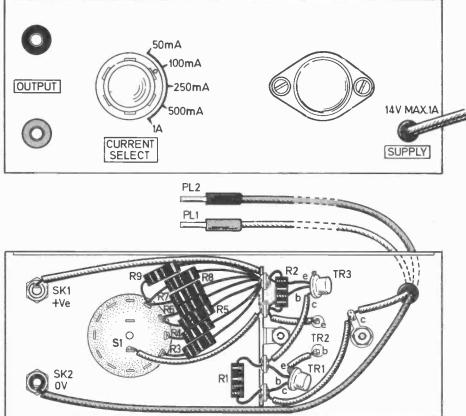
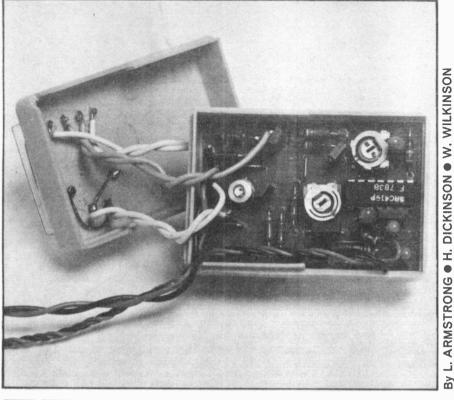


Fig. 2. Layout of components, interwiring and front panel details. Note that S1 consists of one section of a 2 pole-6 way rotary switch with two positions unused.

> All components are mounted on the rear of the front panel. The range resistors are wired directly between the range switch (S1) and the small tag strip.



FE RADIO CONTROL SYSTEM PROPORTIONAL SPEED CONTROLLER

THE conversion of electric signals into mechanical action is the main purpose of radio control systems. This month we deal with a speed controller as applicable to model boats and cars.

PROPORTIONAL SPEED CONTROLLER

A proportional speed controller is a variation of a servo, in this case without the gearbox and with the position feedback potentiometer removed. Once again the ZN419CE integrated circuit is used. As well as these two physical differences there are two other changes, this time in characteristics of operation.

LARGER DEADBAND

As mentioned in the servo section Fig. 4.3 showed the relationship of error to pulse expansion, which in the case of the servo was a narrow deadband and a very steep pulse expansion slope.

If Fig. 5.1 is now referred to the characteristics for the speed controller can be seen. Here a larger deadband is used together with a drastic reduction in the pulse expansion slope to the extent of full drive output being reached very close to full stick-movement (maximum error) which will be equivalent to maximum revolutions on an electric motor.

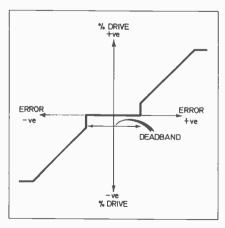


Fig. 5.1. Proportional speed controller error and output drive relationship.

PULSE EXPANSION RELATIONSHIP

Fig. 5.2 clearly shows the pulse expansion relationship for four different input pulse widths.

Case A is when the input pulse is at its minimum value of 1ms. The error between this and the fixed internally generated pulse (1.5ms) is 0.5ms which produces an output drive pulse of 20ms (full drive).

Case B has an error of 0.25ms which produces an output pulse of 10 ms (half-drive).

These two cases are for the input pulse widths less than the internal pulse, producing "positive" drive output pulses. As the error approaches zero so the output pulse will reduce in a linear fashion.

For cases D and C this is where the error is "positive", that means input pulse larger than internal 1.5ms, which will produce a "negative" drive output. (The terms "positive" and "negative" refer to the direction of rotation of the motor.)

It can therefore be seen that with the transmitter stick at neutral (producing a control pulse width of 1.5ms) there will be no error pulse and so the motor will remain stationary. Any change of stick position will produce a pulse width error in the speed controller causing an output drive pulse to be produced and the motor to turn. The larger the stick movement the greater the error and the faster the motor will turn.

TWO IMPROVEMENTS

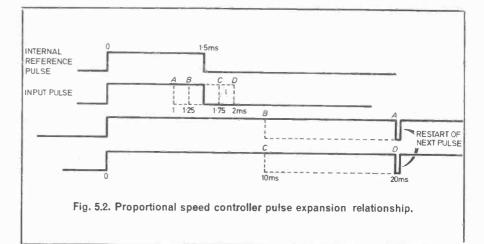
This is the theory of operation of a basic speed controller. The design incorporated in the EE Radio Control System has two improvements over such a basic system.

A relay is used to reverse the direction of rotation of the motor to avoid too many expensive power transistors when using larger motors.

The power gain stage between the servo amplifier and the motor enables external batteries to be used so that the normal receiving batteries are not "flattened" by the motor.

CIRCUIT DESCRIPTION

Fig. 5.3 shows the full circuit diagram of the speed controller. The deadband is set by the capacitor C3. As can be seen this capacitor is much larger than that on the normal servo



amplifier. -C4, R7 and VR2 combine to produce the desired expansion characteristics, VR2 enables the expansion to be adjusted to produce the optimum performance.

The positive and negative drive pulses are combined via R6, R5, TR3, and TR4 to produce a common drive pulse.

TR6, TR7 and R10 from a power stage to give the speed controller the ability to drive larger motors.

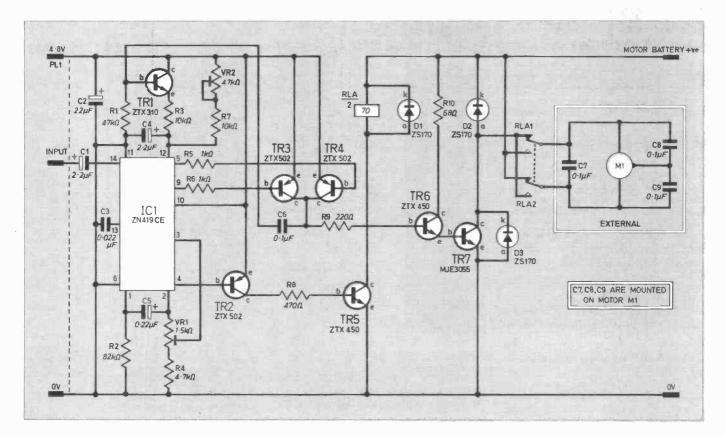
FLIP FLOP AND RELAY

The design of the ZN419CE i.c. includes a flip flop output on pin 4. This is high when the output drive is in one direction and low when in the other. This signal is used to drive the relay RLA which controls the direction of rotation of the motor. TR2, R8 and TR5 are the associated discrete components.

The combination C6, R1, TR1, R3 is designed to ensure that the minimum output pulse width is close to zero other than 2ms as in the normal servo. This ensures that there is a smooth take-off of the motor at low speeds.

The remaining components R2, C5, VR1 and R4 are associated with the internal monostable, in IC1.

Fig. 5.3. Circuit of the EE Radio Control System Proportional Speed Controller. Motor M1 and associated capacitors are external to speed controller.



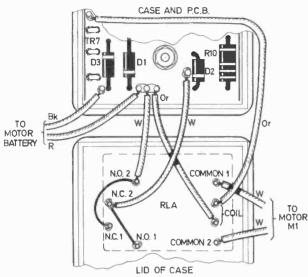
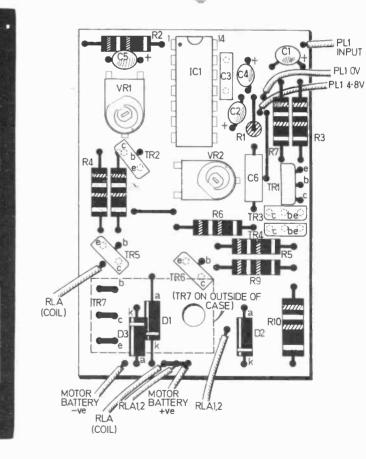


Fig. 5.6. Connections between the p.c.b. and TR7 and RLA, which are mounted on the bottom and top of the case, respectively, see photo (right.)

TO MOTOR M1

.



PROPORTIONA

SPEED CON



Fig. 5.5. Top view of the p.c.b. with all components in situ.



The speed controller is housed in a plastics case of same size as that used for the receiver. The p.c.b. also is comparable in size to the receiver p.c.b. Thus the construction should not prove any more difficult than the receiver-and certainly less "tricky" than the servo.

The power transistor and its heat sink are mounted externally on the bottom of the case. The relay is also mounted outside but on the top of the case.

CIRCUIT BOARD

The p.c.b. pattern is shown in Fig. 5.4. The top side with components in position appears in Fig. 5.5.

Ensure that the i.c., diodes, transistors and polarised capacitors are inserted the correct way round.

The only difficult part is the power transistor TR7. This is mounted outside the case on a heat sink with the leads bent to go through the case and the board. Here they are soldered to the three pins shown. The bolt for the transistor also holds the p.c.b. in place inside the plastic case.

Before fixing the power transistor in position it is wise to check and double check all soldered joints ensuring that no solder bridges adjacent tracks on the p.c.b.

The relay RLA is mounted to the top of the case, Fig. 5.6.

COMPONENTS SPEED CONTROLLER

Resist	ors		
R1	47kΩ	R6	1kΩ
R2	82kΩ	R7	10kΩ
R3	10kΩ	R8	470Ω
R4	4·7kΩ	R9	220Ω
R5	1kΩ	R10	68Ω 1 W

All 1 W carbon ± 5%, except R10

Potentiometers

- VR1 1.5kΩ horizontal mounting miniature skeleton preset
- VR2 47kΩ horizontal mounting miniature skeleton preset

Capacitors

C1	2.2µF	tantalun	n bead	10 V
<u></u>	00.004	a a dia di suana	hood	C 01/

C2	2245	lar	laium	beau	0.	ΣV
C2	0.000	E	ooran	Nic		

00	0	026/01	Ceranni	0
C1	0.	2 uFta	ntalum	head 10V

- C4 C5 0.22 µF tantalum bead 35 V
- C6 0.1μ F ceramic disc
- C7 0.1µF ceramic disc
- Č8 0.1 µF ceramic disc
- C9 0.1μ F ceramic disc

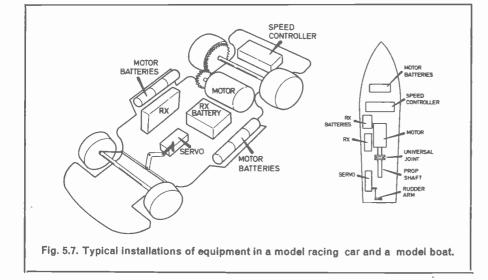
Semiconductors

TR1	ZTX310 <i>npn</i> silicon
TR2	ZTX502 pnp silicon
TR3	ZTX502 pnp silicon
TR4	ZTX502 pnp silicon
TR5	ZTX450 npn silicon
TR6	ZTX450 npn silicon
TR7	MJE3055 npn silicon (plastic)
D1-3	ZS170 1A diode 100V working
	(3 off)
101	

IC1 ZN419CE (SRC419)

Miscellaneous

PL1 3-pin plug (SLM) RLA 6V d.c. 70Ω coil, 2 pole c.o. contacts 2A. (RS 349-181) Case, 60 \times 40 \times 22mm approx. (SLM); Printed circuit board; Small plastic power transistor heatsink; 1 inch × 4BA bolt and nut; Wire for leads; Heat shrinkable plastic sleeving.



EXTERNAL LEADS

External leads should be soldered to the p.c.b. before it is secured within the case:

(a) The three input leads (approximiately 8 inches in length) should be twisted together and soldered to a 3 pin SLM plug, as shown in Fig. 5.5.

(b) A pair of leads for the motor battery (black and red).

Inter-connections between the p.c.b. and the top half of the case can now be made. A pair of leads must be provided to go to the drive motor (see Fig. 5.6).

When and only when you are satisfied that the speed controller is correctly constructed then you can proceed to the next step, that of setting it up.

SETTING UP PROCEDURE

Before plugging in to the receiver, and with the motor disconnected, set VR2 fully anticlockwise and VR1 to approximately the centre position.

Plug in to the receiver output and connect the external 6V motor supply (check correct polarity).

With the transmitter and receiver switched on and all sticks in neutral, movement of VR1 should cause the relay RLA to pull in at one end and drop out at the other end. The correct setting is where the relay is "just" off.

Now connect up the motor.

With the transmitter stick in the neutral position the motor should not turn. Move the stick in any direction the motor should start to turn slowly at first and faster for more movement of the stick.

With the stick in its maximum position adjust VR2 slowly until no further increase in speed is noticed. This is the correct position.

Now check that when the stick is moved in the opposite direction the relay changes over and the motor turns the other way.

It is suggested that when using the speed controller in a boat or car the control stick used is in place of the elevator (see Fig. 2.7 in Part 2).

If, when the system is installed, the motor turns the opposite way to that you require, reverse the connections to the motor or change over the stick plug in the transmitter. DO NOT change round the drive battery.

INSTALLATION

Typical arrangements of equipment including the servo and the speed controller inside a model boat and a model car are shown in Fig. 5.7.

Next Month: Battery charger and fault finding chart.



This new decade will see a revolution in radio communication for ships at sea with the formation of INMARSAT, the International Maritime Satellite Organisation, whose first director general, Mr. Olof Lundberg from Sweden's maritime telecommunications organisation Televerket, set up INMARSAT's London headquarters recently. INMARSAT SAT is a consortium of twenty-nine maritime countries in which Britain with 9.95 per cent of the shareholding has the third largest stake, exceeded only by the USA with 23.5 per cent and the USSR with 14.17 per cent.

The consortium plans to put up six satellites with pairs placed in geostationary orbit 22,300 miles up above each of the ocean zones Atlantic, Pacific and Indian to provide ample channel capacity for telex, telephony, facsimile and data and recorded information communications for a satooms-equipped ship population expected to reach at least 2,000 by 1990; channels will also be available for distress and safety traffic.

This 3 plus 3 configuration will comprise three MARECS satellites supplied by the European Space Agency plus special maritime transponders on three INTELSAT satellites of the International Telecommunications Satellite Organisation. The first MARECS A, is satellite, scheduled for launching into orbit over the Atlantic before the end of this year, from Kourou in French Guiana, on an Ariane LO4 rocket.

The first of the INTELSAT trio is expected to be put into position over the Indian Ocean by late 1981 and though no dates have been set for further launchings these should then follow in fairly quick succession to complete the space segment with the Atlantic zone covered by MARECS A and INTELSAT B, the Pacific by MARECS B and C, and the Indian Ocean by INTELSAT A and C.

Ground Stations

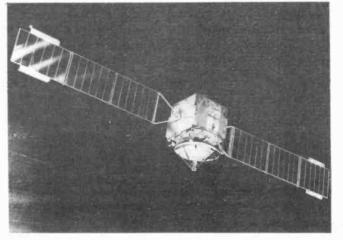
At least a dozen ground stations, and probably several more, will work the six satellites. Britain's will be at Goonhilly where a new 13m diameter dish antenna will beam on the Atlantic, and the Nordic countries will have their terminal, working to the Indian Ocean, at Eik in Norway. at Odessa on the Black Sea and another at Nakhodka near Vladivostok. The station for Greece will probably be at Thermopylae, Italy is thinking of Fucino as a site, Kuwait and Singapore are both planning to have their own, and Australia and Canada will also build ground stations at sites not yet determined. It is expected that other

It is expected that other member countries of IN-MARSAT, notably in South America, Africa, the Indian sub-continent and China, will also set up ground stations. The existing terminals of the US Navy's MARI-SAT satcoms system, at Southbury in Connecticut, Santa Paula in California, and Yamaguchi in Japan will switch over to INMARSAT as MARISAT is phased out over the next two or three years.

With a wide geographical spread of numerous ground stations and double the number of satellites, exclusively dedicated to commercial traffic, INMARSAT will offer a fast high-quality service when it gets off the ground, literally as well as metaphorically.

The British Post Offices Prestel Viewdata system has been purchased by Hong Kong with the service expected to start late in 1980.

The European Space Agency's MARECS satellite, a maritime version of the European Communications Satellite (ECS). MARECS will provide direct telephone and telex links between ships in distant oceans and shore stations.





LOOKING BACK

In March 1930, Baird's much heralded "Televisor" was finally on sale, and the experimental 30-line transmissions from the BBC Brookmans Park station were for the first time accompanied by sound.

Television broadcasting, in fact, had arrived, and the fiftieth anniversary of this milestone is being marked by a special exhibition, entitled *The Great Optical Illusion*, at the Science Museum on March 27 for six months.

ANALYSIS

THE LASER COMES OF AGE

Twenty years ago Dr. T. H. Maiman, a scientist working for Hughes Aircraft Corporation in California achieved light amplification by stimulated emission of radiation which, from its initial letters, generated a new word in the vocabulary of physics—the laser.

Here was an entirely new form of light of great spectral purity, all on one wavelength and with all the waves in a beam in phase with each other. It was coherent light, unlike the light we ordinarily see from the sun or from a tungsten filament which is scattered in a direction and of random wavelengths and energies.

The special characteristic of laser light is that it can contain very concentrated energy and can be in a very fine needlebeam which is so nearly parallel that even at a distance as great as that from the earth to the moon it would still only have a diameter of a couple of miles. In 1960 the laser was rightly hailed as a great invention.

In practice, oscillation is more important than amplification but then the acronym would have become 'loser' and that would hardly do. Nonetheless, in a sense the early lasers were losers. Although recognised as a wonderful technological breakthrough, nobody quite knew what to do with them. The laser was dubbed an invention in search of an application.

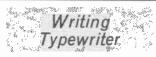
Within months of Dr. Maiman's achievement with his ruby laser, another scientist at the Bell Telephone Laboratories had produced a gas laser and later there came along the semiconductor laser. Today we have a whole range of types each with characteristics which might include exceptional power, or continuous beam or exceptionally directional or particularly coherent. And there are different wavelengths, not all in the visible region, and some offer a choice of colours.

The applications were not long in coming. By 1965 the fine beam of a laser was being used in eye surgery, one of the more significant early applications. Today the laser is found as well in industry, in communications, in research, in measurement, in navigation (the laser gyroscope), in entertainment electronics and particularly in defence.

In 1968 the gas dynamic laser gave a great increase of power which led to speculation on the development of the death-ray of science fiction. The direct-beam laser weapon remains a possibility using newer chemical or electric discharge types. The idea of vaporizing an incoming ballistic missile with its nuclear warhead almost as soon as detected is certainly attractive, and nothing is as fast as a laser beam to make the intercept, travelling at the speed of light. But this is a long way off yet. At present the laser in military use is confined to a secondary role such as range-finding, target marking and weapon guidance, and communications.

Meanwhile more peaceful applications continue to proliferate and it is quite conceivable that at least the semiconductor laser may soon find its place alongside other devices in the electronic hobbyist's kit.

Brian G. Peck



An electronic typewriter that writes the characters with a stylus instead of impacting from a pre-formed metal character has been developed by Centronics Data Corporation. It can "write" up to 20

It can "write" up to 20 characters per second and can be programmed for different styles of script and in different sizes. And because it writes instead of types it is silent in operation.

Named the Quietwriter it is expected to be available early in 1981.

Mobile telephone users in the London area will be able to use direct dialling from their cars instead of being connected by operator. New equipment supplied by Pye and worth £1.75 million will enable 15,000 mobile users to dial direct. -Farnborough '80

This year's Farnborough Air Show (September 1-7), a major electronics as well as aerospace event, will be 30 per cent larger than the last show in 1978.

A Datatext micro-chip based word processing system will be used to process the 100,000 or so admission tickets for invited guests from 100 countries.

Solar Power Down Under

Australia's first high-temperature solar power station is to be located at Whitecliffs, New South Wales. Designed by the Australian

Designed by the Australian National University it uses paraboloidal mirrors to concentrate the solar energy to generate super-heated steam to drive a turbo electric



An order for over 500,000 remote control chips has been awarded to Plessey Semiconductors for use in the French toy manufacturers Joustra's radio control models.

The integrated circuits used are the SL490 transmitter and the ML928 receiver and can also be used for ultrasonic, infra-red or cable transmission.

Originally developed for TV applications, Plessey is designing the i.c.s into toys, TV games and domestic appliances.



Those so puzzling echoes

More than 50 years ago radio engineers and scientists began to investigate a most puzzling and curious phenomenon —the occasional appearance on some h.f. (short wave) transmissions of distinct and quite audible echoes where the second signal was sometimes many seconds or even minutes after the first. Now a radio signal that goes right round the world (as of course happens quite regularly on h.f.) will do so in under a second, while even a signal reflected from the moon will be safely back on earth within about two seconds.

So what could happen to a signal travelling at 300,000km per second that could delay it by several seconds? In other words what path could it take to account for such "long delay echoes" (LDEs).

In 1927 and 1928 a special series of transmissions from Eindhoven, Holland yielded many instances of delays of from 3 to 15 seconds, and even instances of echoes delayed by 3 or 4 minutes. Since then, many investigations have been made and many theories put forward. Yet none has been entirely satisfactory or widely accepted. And there has been a curious change in the reporting of LDEs.

Up to about 1939, there is good evidence that many apparently genuine LDEs were heard and even, in a few but reliable cases, such echoes were recorded photographically or on undulator machines used for machine-telegraphy (this was before the days of the ubiquitous tape recorder). But since 1946 the LDE has become extremely rare.

Some five years ago, Peter Duffett-Smith, G3XJE at the Cavendish Laboratory devised a complex correlation technique to search automatically for LDEs in a noisy background; yet half-amillion transmissions did not yield a single positive result and he concluded that "the phenomenon of LDEs should be treated with reserve".

Certainly it has been shown that some reports were due to deliberate hoaxes, some people feel the echo is often an odd quirk of the brain, and other people have linked them with UFOs. There are also reports that turn out to be due to the practice of overseas relays of h.f. broadcasting stations playing tapes sent out from base slightly out of sync with the home transmission.

Yet those early reports remain convincing, though in many years of listening of h.f., I cannot claim to have heard an LDE. In 1974, a Danish amateur heard a $4 \cdot 6$ -second echo on his "moonbounce" signals on the unusually high frequency of 1296MHz: he believes his signals were being reflected from an ionized cloud near the sun.

Now a Canadian scientist, D. B. Muldrew, has published a long and detailed paper surveying the whole history of LDEs

By Pat Hawker, G3VA

and has.advanced some theories of his own, including the possibility that one of the basic mechanisms is non-linearity in the ionosphere (the old "Luxembourg" effect). This seems to tie several loose ends together.

I have always had the feeling that the key to the puzzle must be in the difference between the pre-1939 and post-1947 conditions. And remember that it was in those years that a vast number of highpower radar, television and megawatt broadcasting stations began to appear.

It is known that powerful signals raise significantly the electron-temperature of the ionized layers and so bring about non-linearity. Exactly why that should affect the non-appearance of LDEs I have no idea—but (at least to me) it seems as plausible a theory as many of the othersI

On the amateur bands

Much that is reported about amateur radio operating seems to place undue emphasis on the unusual contacts: those often fleeting and perfunctory "contacts" with DXpeditions, rare-country or rare-prefix stations or in the all-toonumerous "contests". While undoubtedly all these have an established place within the large framework of amateur radio, personally it has always seemed to me that as equally an interesting and as varied a side is represented by the longer, more leisurely contacts with old friends or the random new ones that come from looking around the bands and responding to CQ calls.

In this way one is likely to find oneself in contact with all those mysterioussounding Russian towns, from Kirov to Kharkov, from Krasnodar to Gorky, from Baku to Tomsk or with the many Scandinavians spread out from the Lofoten Islands to Stockholm, Oslo to Pitea and then, as the "skip" lengthens. with friendly amateurs down in Rio or in remote Anchorage, Alaska or busy New York. Even with a simple transmitter one can take advantage of those powerful American 1kW, "6-element Yagi, up 70ft" stations spread right across the United States, or perhaps look into Europe through the ears and fingers of the small 3-watt morse transceivers of the Heathkit HW7/HW8 class used so many of the QRP (low power) bv enthusiasts.

Or you come across the chap who keeps an accurate cross-index to his log and can immediately tell you he last "worked" you in August 1947 or the amateur who recalls his visit to London or his hopes of coming soon. Perhaps he is particularly proud of his "homebrew" equipment, or that he has been licensed for just three weeks—or perhaps 50 years.

The other day Nick Carter, G2NJ, an old-timer, recalled to me the early 1920s and his memories of the days when amateurs in the UK were allowed to operate on 440 metres and had their own "signature tunes" (for many years amateur stations were permitted to play one gramophone record for "test" purposes) such as the Skater's Waltz (G2UV still an active amateur) and "A pair of sparkling eyes" (G2MO).

There are all those names in the log-Vlad, Igor, Jirka, Adam, Gerard, Zoli, Bengt, Czeslaw, Leif-and the pleasure when one suddenly comes across a callsign that strikes a distant chord and turns out to be old Stan in Swansea with whom one remembers sharing Hut 4 in 1942 but has not met for over 35 years.

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This type of amateur activity has changed remarkably little over the years. Of course, modern equipment is easier, far easier, to operate with its voice or key operated changeover switching, its vastly improved stability, its more penetrating s.s.b. than the old a.m. equipment. But the pleasure to be derived from such contacts has changed so very little.

In the UK, amateur radio has been growing rapidly in numbers, with some 26,981 licences current at the end of 1979 and with 3155 new licences issued by the Home Office during the year.

Droitwich to move?

One of the lesser noticed changes that will stem from the World Administrative Radio Conference at Geneva last year is a small frequency shift made to the longwave broadcasting band. So small that it will hardly be noticed by the average listener when it occurs some time around 1986–90. But for many enthusiasts this change poses a particular problem.

The very high stability of the BBC Droitwich station, on exactly 200kHz, has for long made it an ideal "frequency standard" particularly when all modulation is stripped off by passing the carrier through a crystal filter and then generating harmonics right up to h.f./v.h.f. as calibration markers. Far better and more reliable (except when the Christmas gales damaged the aerial) than the WWVtype of h.f. standard frequency transmissions (5, 10, 15MHz etc).

But it looks as though Droitwich will have to move to 198kHz, an exact multiple of the 9kHz channel spacing. Of course, with modern digital techniques one can divide by any integral—but, well, how can one get enthusiastic about the number 198 compared with that nice conveniently rounded 2001 So make the best use of Droitwich while you can.

CB in the States

At a time when CB appears to have been postponed indefinitely in the UK, it is interesting to learn that the Federal Communications Commission are proposing to relax still further some of the current CB rules in the United States. On December 19, 1979 at an open meeting of the FCC, preliminary steps were taken to consider allocating more channels to s.s.b. and to explore ways of making it legal for CB stations to work over longer distances and to use variable frequency oscillators.

However, the FCC is apparently considering a suggestion that before such relaxation of the rules, operators would have to demonstrate their knowledge of FCC rules which are seldom obeyed by some CB operators.

$\begin{array}{ c c c c c c c c c c c c c c c c c c c$	LINEAR LF356 80p NE531 98p THIS IS ONLY M308 60p NE555 23p THIS IS ONLY M308 60p NE556 60p A SELECTION M308 60p NE567 100p 709 35p LM324 45p RX76477 230p 741 16p LM379 410p TBA8105 100p 748 30p LM390 75p TDA1022 620p 7106 850p LM3900 65p TL084 125p CA3046 55p LM3911 100p ZN425E 390p CA3046 55p M57160 59p ZN425E 390p
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9

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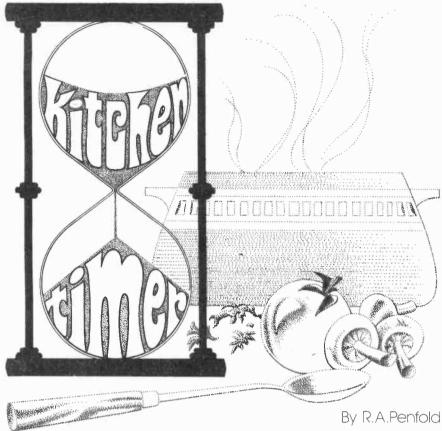
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HIS unit is primarily intended for use as a kitchen timer, but it could, no doubt, be employed in other timing applications. The timer sounds an audio alarm at some predetermined time after the unit is switched on, and the times available range from 1 minute to 2 hours in 1 minute steps.

By R.A.Penfold

The required time is selected by adjusting two switches, and once the alarm has sounded the unit is silenced simply by switching off using the on/off switch. In order to start another timing run it is only necessary to switch the unit on again, having readjusted the time setting if necessary.

Fig. 1. Complete Circuit of the Kitchen Timer.

METHOD OF OPERATION

Most electronic timers are of the CR type where a capacitor is charged via a resistor. The voltage across the capacitor rises until a certain trigger level is achieved, and then the circuit switches on an alarm of some kind.

Such timers can be very accurate where only fairly short times are required, say no more than a few minutes, but they tend to be unreliable where longer times are called for. This is because in a practical circuit the length of the timing period is usually calculated from the formula:

t = CR

where t is time (seconds)

C is capacitance (farads) R is resistance (ohms)

For a one second delay this would

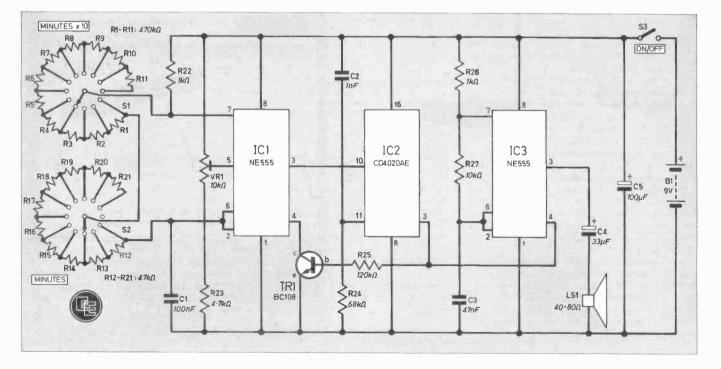
mean values of 1 microfarad for C and 1 megohm for R.

In order to obtain a delay of a couple of hours (7,200 seconds) timing component values of something like 1,000 microfarad and 7.2 megohm would be required (1,000 x $7 \cdot 2 = 7,200$).

Values of this order are readily available, but it would be necessary to use an electrolytic timing capacitor due to the high value needed, and such capacitors tend to have high leakage levels.

This would result in a significant amount of the charge current leaking away during the charge period, extending the charge time and possibly preventing the trigger voltage being reached at all.

A more sophisticated form of circuit is needed in order to give good



predictability of the timing period plus consistent and reliable performance.

THE CIRCUIT

The usual form of precision long timer circuit has a *CR* oscillator and a counter circuit. When the counter has received a certain number of input pulses from the oscillator it operates the alarm.

The counter circuit usually has a high division rate (1,000 or more) so that the oscillator can operate at a comparatively high frequency and still provide long timing periods. This enables quite low value timing components to be used in the oscillator.

Consequently there is no need for the timing capacitor to be an electrolytic component and good results can therefore be produced.

CLOCK OSCILLATOR

The circuit diagram of the Kitchen Timer unit is shown in Fig. 1, and this is of the type outlined above. The well known 555 timer i.c. is used here in the astable mode to provide the oscillator section of the timer (IC1). This type of circuit is normally termed the "clock oscillator" or just the "clock".

This part of the circuit operates with Cl first charging up to twothirds of the supply voltage via R22 and the resistance between pins 6 and 7 of ICl and then discharging through the resistance between pins 6 and 7 and an internal transistor inside ICl until it has a charge equal to one-third of the supply potential. It then commences charging to twothirds of the supply voltage again, and the circuit continuously oscillates in this manner.

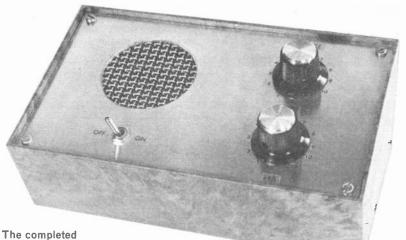
TIMER RANGE

The value of one timing element must be made variable in order to give a range of times, and the resistance between pins 6 and 7 of IC1 has been made the variable element.

Times of 0 to 10 minutes can be provided by S2 and R12 to R21. With S2 adjusted so that only one resistor is connected into circuit, the oscillator will run at a frequency that gives a one minute delay.

Switching two resistors into circuit will give double the timing resistance, half the previous frequency, and hence double the time delay. Three resistors will give three times the delay and so on.

The switch S1 can be used to switch further timing resistors into circuit (R1 to R11), and as these have ten times the value of the other timing resistors they will each increase the delay by ten minutes when switched into circuit.



Kitchen Timer showing front panel layout.

It has been assumed here that the delay is proportional to the resistance between pins 6 and 7 of ICl but actually the resistance of R22 must also be taken into account. However, since R22 is only effective while Cl is charging, and its resistance is always low in comparison to the main timing resistance, any inaccuracies this introduces will be negligible.

TIME PERIOD ADJUSTMENT

Resistors VR1 and R23 shunt the internal potential divider of IC1 that sets the discharge threshold for C1, and VR1 can be used to trim the clock to the correct frequency range so that the appropriate delay times are produced.

Component tolerances and similar considerations would otherwise give a relatively low level of performance. Raising the discharge threshold voltage of C1 reduces the clock speed since it will obviously take longer for C1 to charge to this level. Reducing the discharge threshold voltage has the opposite effect.

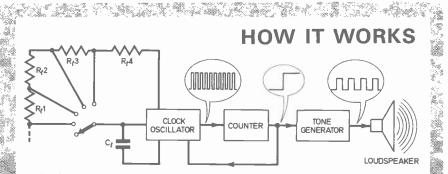
COUNTER

The CMOS chip IC2 is the basis of the counter circuit and this has 14 divide-by-2 stages connected in series giving a total division rate of 16,384. Components C2 and R24 provide a positive pulse to the reset terminal of IC2 at switch on so that it is automatically reset to zero and the output pin 3 assumes a low voltage.

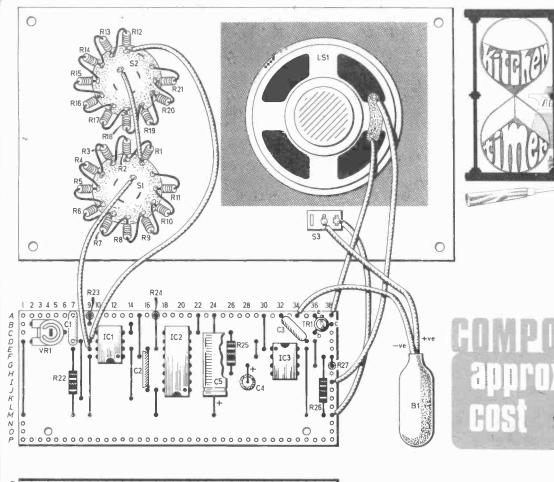
If allowed to continuously cycle, after 8,192 input pulses the output would go from a low voltage to virtually the full positive supply voltage. After another 8,192 input pulses it would return to a low voltage, and so on, with 16,384 input pulses producing one complete output cycle.

This does not occur here because when after the first 8,192 input pulses the output goes high, TR1 will be switched on by the base current it receives through R25; TR1 then ties IC1 pin 4 to the negative supply and this disables the clock oscillator.

The circuit therefore latches with the output of IC2 in the high state until the power is switched off.



A clock oscillator operates at a frequency determined by a capacitor C_t and a network of switched resistors R_t1 , R_t2 , R_t3 , R_t4 ... (providing a wide range of frequencies and times). The output of the counter circuit is normally at a low voltage, but after it has received 8192 pulses from the clock oscillator its output goes high. This switches on a tone generator oscillator which drives a loudspeaker and provides an audible alarm. The counter's high output is also used to disenable the clock oscillator, thus preventing further pulses being applied to the counter (which would cause its output to go low again after a further 8192 clock pulses).



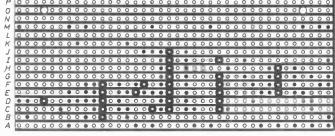
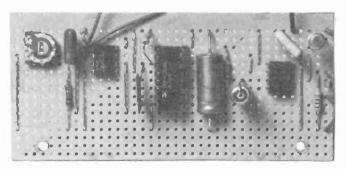
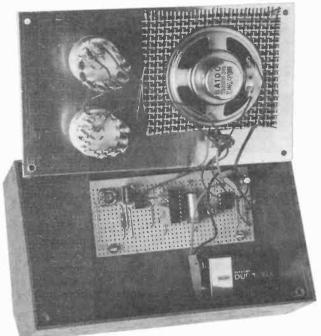


Fig. 2. Circuit board component layout and rear view of front panel showing interwiring details. Also shown is circuit board underside and breaks in the copper strips.



Components mounted on the finished circuit board. Take particular care to ensure all link wires are correctly positioned.



The completed timer with front panel displaced showing positioning of circuit board. The range resistors are mounted directly on the selector switches.

E10

TONE GENERATOR

Another 555 timer chip IC3 again connected in the astable model is used as the basis of the alarm generator. Timing components R26, R27 and C3 produce an operating frequency of about 1kHz, and the output from pin 3 is supplied to a high impedance speaker via d.c. blocking capacitor C4.

Pin 4 of IC3 is normally held at a low voltage by the output of IC2, but this is taken high at the end of the timing period, enabling IC3 to function properly and causing the audio alarm to be generated.

Capacitor C5 is a supply decoupling component. The operating frequency of a 555 astable circuit is not significantly affected by variations in supply voltage, and so a non-stabilised 9 volt battery supply can be used.

The current consumption of the unit is about 15mA under quiescent conditions, and about 50mA or so when the alarm is operating. This gives many hours of operation from a PP6 battery.



CIRCUIT BOARD

Most of the circuit is constructed on a piece of 0.1 inch matrix stripboard, 16 strips by 38 holes, using the layout illustrated in Fig. 2. Drill the two mounting holes and make the 27 breaks in the copper strips before soldering the components and link wires into position. Note that IC3 has the opposite orientation to IC2 and IC1.

As IC2 is a cmos device, it can therefore be damaged by high static voltages. It is therefore recommended that a socket should be used for this component and that it should not be plugged into position until the unit is otherwise complete.

This device will probably be supplied in some form of protective

	COMPONENT	S	R
Resistors R1-R11 R12-R21 R22 R23 All ╁ W c	470kΩ 2% tolerance (11 off) 47kΩ 2% tolerance (10 off) 1kΩ $4 \cdot 7kΩ$ arbon ± 5% unless otherwise st	R24 R25 R26 R27 tated	68kΩ 120kΩ 1kΩ 10kΩ
Potentiome VR1	ete r 10kΩ subminiature horizontal pi	reset	
Capacitors C1 C2 C3	100nF polyester 1nF ceramic 33μF 16V elect.	C4 C5	47nF polyester 100µF 10V elect.
Semicondu IC1, IC3 IC2 TR1	u ctors NE555 timer i.c. (2 off) CD4020AE CMOS 14-bit binary BC108 <i>npn</i> silicon	counter	Shop Talk
38 holes or solder	ous 1-pole, 12-way rotary switch 1-pole, 11-way rotary switch (12 miniature s.p.s.t. toggle miniature loudspeaker, impedan 9V PP6 type front box 162 × 97 × 68mm, or × 16 strips; two control knobs; con pins; speaker covering; mou tecting wire.	nce 40-80 o r similar; (battery cor	hms 0·1 inch matrix stripboard nnector; 16 pin d.i.l. socket

packaging and it should not be removed from this packaging until it is to be plugged into circuit.

CASE

A sloping front case having approximate dimensions of $162 \times 97 \times 62$ mm (maximum height) makes a good case for this project, but any case of about the same size should be equally satisfactory.

The switches SI and S2 are mounted one above the other on the right hand side of the front panel, and the loudspeaker plus S3 are situated opposite these on the left hand side.

A cut out about 45mm in diameter must be made for the loudspeaker and some speaker cloth or fret is glued in place behind this. Very few miniature loudspeakers have provision for screw fixing and it will almost certainly be necessary to glue the speaker in place, taking care not to smear adhesive onto its diaphragm.

FINAL ASSEMBLY

The component panel is mounted on the base panel of the cabinet on the left hand side and well towards the front, leaving a space for the battery behind the component panel. However, before finally mounting the component board it is necessary to complete the remaining wiring which is shown in Fig. 2. This mainly consists of wiring the timing resistors to S1 and S2. These resistors should not be allowed to project too far back from the range switches or they will prevent the front panel assembly from being slotted into the rest of the case.

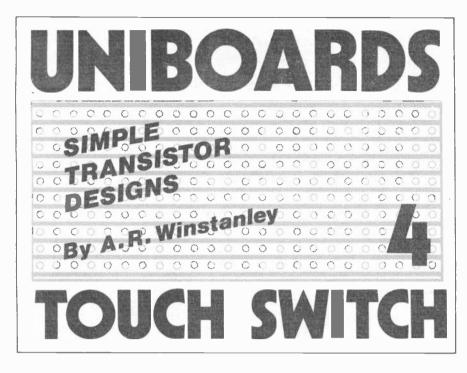
ADJUSTMENT

Initially VR1 is adjusted so that its wiper is at roughly the centre of its track. With both S1 and S2 set fully anticlockwise so that none of the timing resistors are connected into circuit, the alarm should be activated within about one second of the unit being switched on. If it is not, switch the unit off at once and check for wiring errors.

With S2 advanced one position there should be a delay of very approximately one minute between switch on and the alarm sounding. There will almost certainly be a significant error though, and this must be minimised by trial and error adjustment of VR1. Clockwise adjustment of VR1 will increase the delay and anticlockwise adjustment has the opposite effect.

When the one minute delay is approximately correct, set the range switches for a somewhat longer delay and make the final fine adjustments to VR1 to obtain a suitable level of accuracy, again using trial and error.

Ц



A TOUCH operated switch is a useful and popular example of the application of electronics and this article describes the construction of such a device.

In brief, a relay is made to switch on or off by simply touching the appropriate touch pads on the front panel of the Touch Switch unit. The circuit detects the resistance of the skin across the pads and then operates accordingly.

Any load, mains or otherwise, may be switched provided that the electrical specifications of the relay contacts are adhered to. The unit to be described here was designed to be powered from the 9 Volt Power Power project described last month.

CIRCUIT DESCRIPTION

The circuit diagram of the Touch Switch appears in Fig. 1. Transistors TR1 and TR2 are special types of transistors called "Darlington transistors". They have the usual three terminals but internally they actually incorporate two individual transistors as the circuit symbol illustrates. The major advantage of the Darlington is the superior gain parameter obtained from the use of two transistors: gains of 5,000 to 25,000 are not uncommon. A "normal" bipolar transistor may have a gain of several hundred only.

The circuit operates as follows. When the oN contacts are bridged with a finger, the base of TR2 is connected to the positive supply line through R3, R4 and the resistance of the skin. Base current (although very small) therefore flows and TR2 conducts, causing RLA to energise (relay contacts close) and D1 to illuminate. Without TR3, if the finger is removed from the oN pads then TR2 would switch off and the relay contacts open; TR3 has been incorporated to act as a latching transistor. When TR2 is conductive, TR3 is also biased into operation. Current therefore flows through R7 and supplies base current to TR2, thereby keeping this device switched on.

If the finger is removed from the on contacts therefore, TR3 and TR2 will remain conductive, latching the circuit and ensuring that the relay remains energised (contacts closed).

If the oFF pads are now touched, base current for TR1 flows through the skin and R1. This switches TR1 fully on. The base current for TR2 is therefore diverted to ground. This cuts off both TR2 and TR3, and so the relay switches off. If the finger is removed from the oFF touch contacts then TR1 will cease to conduct with the relay remaining in the oFF state. The circuit is so sensitive that even a resistor of 30 megohms placed across the pads will operate the circuit. (This implies a base current of less than one microamp!).

Any a.c. signal which the human body may present to the very sensitive base circuits when the pads are touched is removed by C1 and C2. They also remove the possibility of relay chatter if the pads are touched only very lightly. Capacitors C3 and C4 serve to decouple the power supply.

The reverse-connected diode D2 shorts away any back e.m.f. generated when the relay coil switches out quickly, which might otherwise damage TR2.

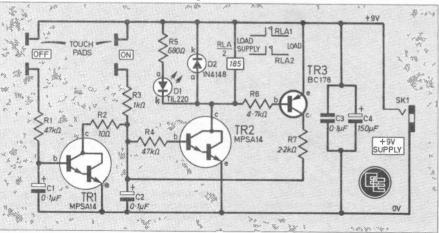
The desired load is controlled through both sets of relay contacts, RLA1 and RLA2. The circuit diagram supposes that a mains load will be driven. The relay chosen should have contacts suitable to comfortably handle the load connected.

Finally, the 9 Volt Power Supply is connected to the Touch Switch unit via SK1. The power supply requirements are 9 to 15V d.c. at approximately 50mA. As a suggested application, the touch switch could therefore be wired to operate successfully in the car as well as in the home.



The complete Touch Switch is constructed on a piece of 0.1 inch matrix stripboard, 10 strips \times 24 holes, as depicted in Figure 2. Two 6BA clearance holes are drilled in the locations shown. These holes will permit the support of the completed stripboard with the appropriate mounting hardware. Eleven breaks are required in the copper strips, and these can be

Fig. 1. Circuit diagram of the Touch Switch.



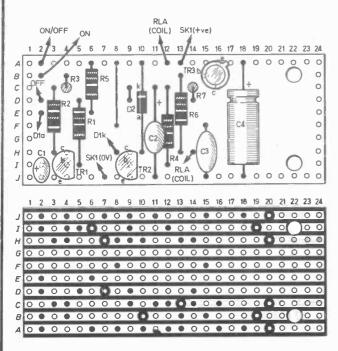


Fig. 2. Layout of the components on the topside of the board, showing wiring connections, and breaks necessary on the underside of the board

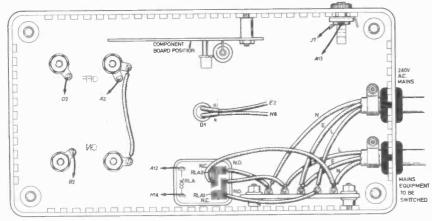
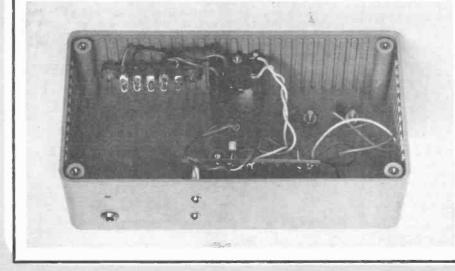
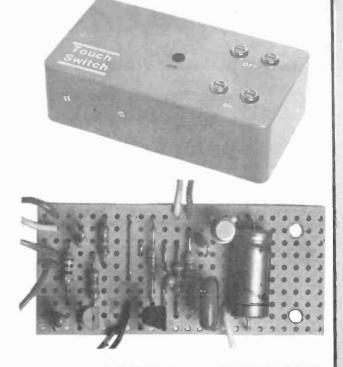


Fig. 3. Complete interwiring details between component board and case mounted components.





COMPONENTS
$\begin{array}{c} \text{Resistors} \\ \text{R1} & 47 k \Omega \\ \text{R2} & 10 \Omega \\ \text{R3} & 1 k \Omega \\ \text{R4} & 47 k \Omega \\ \text{R5} & 680 \Omega \\ \text{R6} & 4 \cdot 7 k \Omega \\ \text{R6} & 4 \cdot 7 k \Omega \\ \text{R7} & 2 \cdot 2 k \Omega \\ \text{All } \frac{1}{2} \text{W carbon } \pm 5\% \end{array}$
$\begin{array}{llllllllllllllllllllllllllllllllllll$
Semiconductors TR1, 2 MPSA14 Darlington silicon <i>npn</i> (2 off) TR3 BC178 silicon <i>npn</i> D1 TIL220 red light emitting diode D2 1N4148 silicon small signal diode
Miscellaneous RLA 185 ohm 12V coil and at least two sets of normally open contacts rated to suit applied load SK1 3.5mm jack socket
Stripboard: 0.1 inch matrix size 10 strips × 24 holes; plastic case type BIM 2005/15 or similar (150 × 80 × 50mm); countersunk 4BA bolts/ nuts/solder tags/cups for touch pads; 6BA mounting hardware, nuts/ bolts/spacers; mounting clip for D1; mains cable rated to suit application; rubber grommets to suit mains cable; 5-way tag strip; cable grips for mains cable.
Approx. cost Guidance only £6.00 excluding case

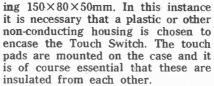
made with a "spot face cutter" or a hand-held twist drill.

It is important that miniature components are used in this design, for it will be seen that the component arrangement on the circuit board is very compact. In this respect, ¹₄watt resistors must be employed, and tantalum bead capacitors have been selected for C1 and C2 because of their small size.

Provided that the diagram is followed carefully there should be no problems, but, as usual, care should be exercised regarding the soldering of the semiconductors. In particular, the tantalum capacitors must be soldered in the right way, as must the diode D2. Note the orientation of the transistors.

ASSEMBLY

The prototype unit was housed in a grey Bimbox type 2005/15, measur-



The construction of the actual touch pads on the prototype consisted of a 4BA countersunk screw with a screw cup placed under the head to give a neater appearance. Connection to the touch contact is by means of a solder tag under the mounting nut.

WIRING-UP

The complete interwiring is shown in Fig. 3. Stranded lightweight interconnecting wire can be used throughout, with the exception that wiring at mains voltages should be suitably rated (3A minimum). All soldered joints which are at mains voltage must be of a good quality. The mains (or other) supply enters and leaves the case via holes fitted with grommets in the case end. The cables should be fitted with grippers for safety reasons. A 5-way tag-strip was found to be a convenient interface between the cable and relay contact tags.

The l.e.d. can be mounted on the front panel using the special plastic clip normally provided with it; the relay can be stuck down with doublesided adhesive foam strip.

With construction complete, connect up the 9 Volt Power Supply, or other suitable supply (9 volt to 15 volt) and then switch on. Touch the oN pads: the relay should be heard to click into operation and the l.e.d. should illuminate. Touching the orr pads should cancel the relay and extinguish the l.e.d.

The Touch Switch is then complete and ready for use.

Next Month: Audio Tone Generator



IBA TECHNICAL REVIEW NUMBER 12— TECHNIQUES FOR DIGITAL TV Editor C. W. B. Reis Price £1.50 Size 225 × 195mm 72 pages Publisher Independent Broadcasting Authority ISBN 0 308 423 X

 A^s the title suggests, this is a review of current practices and future trends in digital TV techniques within the IBA and as such requires a high level of understanding by the reader if he is going to be able to get anything out of the material presented. Chapter headings such as "Digital Sub-Nyquist Filters" and "A Low Bit-Rate System for Digital Video"^b give a good idea of the standard of knowledge assumed.

This publication is effectively intended for engineers and students directly involved in the field of broadcasting and as such is likely to have only limited appeal although presentation is clear and precise with a larger number of line drawings and photographs.

NEWNES	BOOK OF AUDIO
Editor	K. G. Jackson
Price	£4.95 Limp
Size	250 × 185mm 144 pages
Publisher ISBN	Newnes Technical Books 0 408 00429 0

S^{EVEN} well-known specialists have contributed articles to this highly readable, well illustrated survey of techniques and equipments currently in vogue in sound reproduction systems. The names of the contributors will be familiar to all hi-fi magazine devotees; this fact should provide sufficient recommendation to others, especially those seeking to acquire their first hi-fi set-up or wishing to up-date an existing system. For them this book will prove a most valuable technical reference, and help them explore the jungle of the market place.

As the title suggests, the contents are not restricted to

the car is recognised by a section which helps the motorist to get the best from in-car entertainment equipment.

"hi-fi" as generally understood. Recording techniques are explained at length, aided and abetted by another section on microphones. The increasing importance of audio in

MECHANICAL WORLD ELECTRICAL YEAR BOOK 1979/80

Editor	R. Warring
Price	£3.95 Paperback
Size	155 × 105mm 383 pages
Publisher	Argus Books
ISBN	0 85242 679 8

T HE composition of this book reveals just how interwoven these two branches of engineering have become.

The purely electronic information is pretty well as comprehensive as you could wish for in a pocket book: colour codes, semiconductor theory, lists of current discrete and integrated devices, circuit theory and formulae, soldering, wire gauges—these are random samples.

The "electrical" information includes resistivities of metals, properties of plastics, insulating materials, electric lighting installation, cables, fuses, motors, thermostats, electroplating, metric/English threads, SI Units and much more.

In brief, a wealth of information clearly laid out in text, tables and diagrams, and always conveniently at hand.

One Armed Bandit October 1979

We offer our apologies for three mistakes that appeared in the layout diagrams of the One Armed Bandit project.

In Fig. 2, underside view, a break is shown at location O35. This is not required and should be bridged with a short piece of wire. Also in Fig. 2, the end of R14



located at O29 should be removed and connected to A26. In Fig. 3 the two connections P64 and E64 from the diode "reels" should be transposed and annotations "D6-D12 and D20-D26" transposed to agree with the circuit dlagram.

Eversure by Anthony John Bassett

THE Prof. is discussing Guitar Fuzz Boxes with Bob, whose friends Tom and Maurice have brought a couple of faulty ones for his attention.

The Extra-

Experiments

Profess

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ordinar

DIODES

"In this circuit (Fig. 1) the signal is amplified before being clipped by means of two germanium diodes (D1, D2). Instead of using a variable resistor in series with the diodes to control the fuzz depth, as in the 'Simple I.C. Fuzz Box' (Fig. 7 last month), the potentiometer VR1 is used as a volume control to determine how much signal is presented to the diodes after amplification by the input stages, and this determines the proportion of the signal which is clipped away.

"Another way in which small signal diodes can be used to produce fuzz effects is by placing them in the feedback loop of an amplifier. Non-linear elements such as diodes, when placed in the feedback loop of an amplifier will result in non-linear amplification and this is often used in fuzz box circuits."

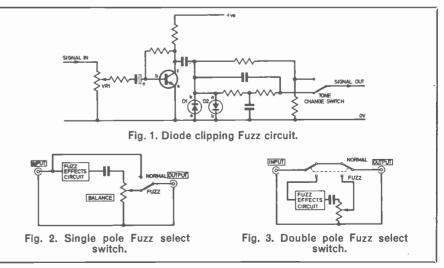
FOOT SWITCHES

"Prof., most fuzz boxes have a footswitch which the guitarist presses to change from 'normal' sound to 'fuzz' sound and back, but I notice the diagrams you have shown do not include the footswitch as part of the diagram. Why is this?" "This is because there are various types of footswitch available and in many instances the footswitch arrangements are interchangeable between the various fuzz circuits.

"The simplest footswitch arrangement in common use uses a singlepole changeover switch (Fig. 2). Here the footswitch connects the output of the fuzz box either to the input or the output of the fuzz circuit, which could be satisfactory in many instances. However, in the 'straight through' or normal sound position of the footswitch the input of the amplifier is still connected to the signal path and this has a slight effect on the sound of the guitar. "Most guitarists prefer a slightly more complicated arrangement using a double-pole footswitch (Fig. 3) and here for 'normal' sound the guitar really is connected 'straight through' the fuzz boxes so that the electronic circuitry for producing fuzz sounds should not affect the sound until 'Fuzz' is selected by operating the switch."

BATTERY SAVER

"In some fuzz boxes the footswitch is arranged to switch the battery supply 'off' for normal playing and 'on' for fuzz. This is a good idea as it means that during normal playing



the circuit does not drain the battery and consequently lasts much longer.

"However, this type of circuit must be carefully designed and built, in order to avoid a loud 'click' when the footswitch is pressed, and possibly a delay between pressing the switch and the appearance of a fuzz effect."

MODERN TREND

"Although this type of problem might be solved with a more complicated switching arrangement, the trend in modern electronics is to replace the moving mechanical parts with electronic circuitry wherever possible, and this could be applied to fuzz boxes with considerable benefits.

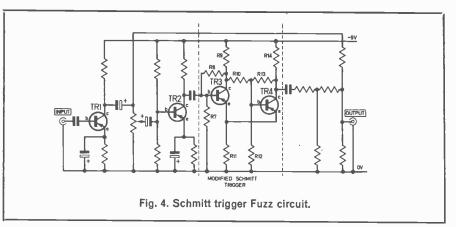
"The mechanical changeover switch could be replaced with a simple single-pole on/off switch which would send a control signal to a logic circuit. The logic circuit could then control a more complex switching arrangement using static electronic switches, with no moving parts to jam, break or become covered in dust or corrosion, and which would route the signal through various effects as required by the player.

"Low-cost CMOS analogue switches such as type 4016 are available at low cost and would be suitable for this purpose. For a little more expense electronic attenuator i.c.s will do even better, especially for the professional user."

SCHMITT TRIGGER FUZZ

"Prof., wouldn't it be possible to use a bistable circuit such as a Schmitt trigger to produce fuzz effects?" enquired Bob, "By feeding the guitar signal through such a circuit it would become totally squared and should give a very intense fuzz effect."

"That is a good idea, Bob, and it has been tried with some success. However, it also produces another effect which, although it is rather novel and interesting may in practice be inconvenient.



"The Schmitt trigger circuit responds not only to the fundamental frequency of any note played, but also to many of its harmonics, usually one at a time in rapid succession. The resulting output from the Schmitt trigger is a series of notes harmonically related to the guitar note being played, in addition to the note itself, and because it is difficult to control is not usually considered to be musically valid.

"By modifying the circuit of the Schmitt trigger this problem can be overcome and an example of this is shown in Fig. 4. Here, TR3 and TR4 form a modified Schmitt trigger circuit, with additional 100 kilohm resistors (R8, R13) from the collector to base of each transistor.

"This moves the two trigger-points of the Schmitt circuit much closer together, so reducing its hysteresis very greatly. As a result, although the output is still a square wave, it is more closely related to the actual note being played and does not show rapid jumps from one harmonic to another.

"In this circuit there is no bias supply to the first transistor, however, like Fig. 1 last month, the circuit does actually work and derives its bias from transistor leakage."

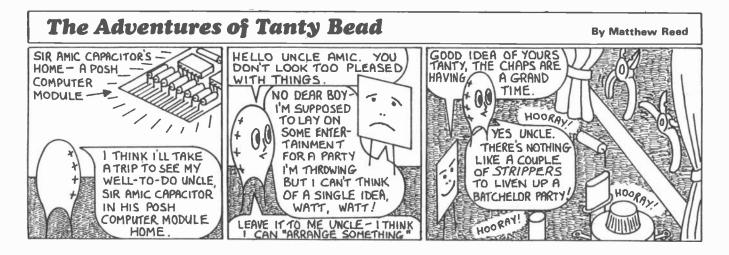
"Prof., I have seen Schmitt trigger circuits which have a 'hysteresis control' for moving the two trigger points closer together or further apart," remarked Bob. "Now I wonder what happens if the hysteresis control is moved to the point where the trigger points coincide ('zero hysteresis') then taken even further, what happens to the trigger points? Do they simply move past one another or annihilate one another? What happens?"

^aAn interesting question, Bob. As the hysteresis points are two very definite points of action of the circuit, one might expect by extrapolation that they would continue to exist even when the control is moved past the point of zero hysteresis, on the other hand one might also logically deduct that as one is a point of action in a positive direction, and the other in a negative direction, they might annihilate each other on contact!

"One could try to find the answer by actually experimenting practically with a Schmitt trigger circuit or by extrapolation and deductive means as a mental exercise!"

Although the Prof. could tell them the answers, he is content to let them find out for themselves whilst he casts a critical eye over the faulty Fuzz boxes—and finds that they use yet more different circuitry for generation of fuzz effects.

To be continued



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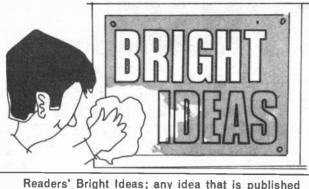
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will be awarded payment according to its merit. The ideas have not been proved by us.

RESIST REMOVER

After etching a home made p.c.b. the constructor is faced with the small problem of removing the etch resist ink from the copper tracks. An inexpensive way of doing this is to give the board a quick squirt with cigarette lighter fluid available in "bubble sachets" for a few pence. The ink can then be wiped off with a cloth. The board requires only a quick rub with a cleaner before soldering. Mr. M. Stansfield,

Mirfield, W. Yorks.

RING PADS FOR P.C.B.

While using etch resist transfers I found that I had run out of pads. As a substitute I used some Reinforcement Rings intended for file paper, available from W. H. Smiths. I stuck these onto the copper side of the board so that the hole for the component lead was in the centre of the ring. Then I filled in the space in the middle with etch resist ink and removed the ring leaving a spot of etch resist ink.

I found that this worked quite well in circuits where the components don't have to be spaced close together like power supply boards and other projects where i.c.s are not used.

> Mr. R. Jonasar, Aylesbury, Bucks.

SHOCK ABSORBER

The normal method of attaching a p.c.b. (or other circuit board) to the project case is by nuts, bolts and spacers. This has the disadvantage that if the boxed project is dropped or jolted, the impact force is transmitted to the p.c.b. components and wiring possibly damaging them. I overcame this by using no spacers, but instead a thin layer of foam plastic between the p.c.b. underside. This then acts as a shock absorber/spacer.

> P. Haddad, Penketh, Cheshire.

PLUG STAND

It is often difficult to solder small audio plugs because both hands usually have to be free to do the work. If the plug is held in one hand it doesn't take long for the plug being soldered to get too hot to hold.

I use a small block of wood with different holes drilled in to solve this problem. The size of the hole(s) to be drilled is the same as the plug pin(s).

Loudspeaker DIN plug holes are made with a 3/64 inch (0.1 mm) drill bit and the flat contact is made with three of these holes drilled close to each other and the wood in between chiselled out.

· + - -

P. Humphrey (aged 14), Mealbank, Cumbria.

SCRATCH RESIST

To aid the drilling of chassis panels ordinary white sticky backed plastic (Fablon) may be employed. The construction lines and drilling points necessary are drawn onto the plastic in pen or pencil and the sheet then applied to the panel. This makes scribing marks unnecessary and at the same time helps to prevent accidental damage during drilling.

> J. Winter, Forfar, Angus.

COMPONENT BASE

When building a project I usually take out all the components I need and put them on the table. But many times I forget where I put them, or on other occasions something falls on the floor and I waste time looking for it. By pressing a piece of plasticine on the table and inserting the components into this, the components are prevented from rolling around on the table and falling to the floor. P. Mallia,

Santa Lucia, Malta.

GROMMET FEET

We have found that small grommets make ideal non-slip feet for calculators and the like, which are prone to slide around when in use on smooth surfaces. The grommets are cut through the central grove, providing two feet which can be stuck onto the base of the unit.

K. R. Nash & A. M. Williams, Swansea.



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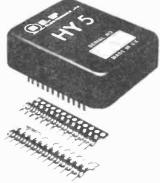
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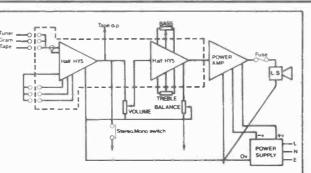
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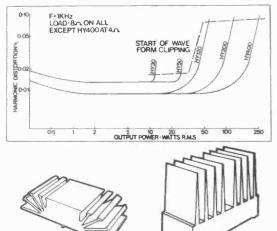


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The HY5 pre-amp is compatible with all I.L.P. amplifiers and P.S.U.'s. It is contained within a single pack 50 x 40 x 15 mm, and provides multifunction equalisation for Magnetic/ Ceramic/Tuner/Mic and Aux (Tape) inputs, all with high overload margins. Active tone control circuits; 500 mV out. Distortion at 1KHz-0.01%. Special strips are provided for connecting external pots and switching systems as required. Two HY5's connect easily in stereo. With easy to follow instructions.

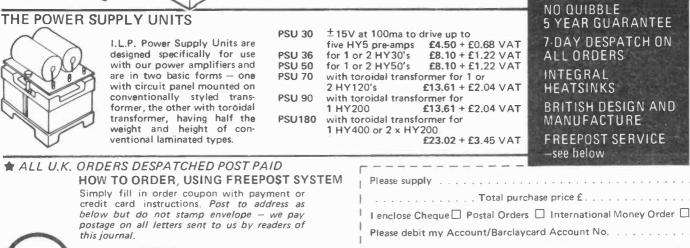
£4.64 + 74p VAT

THE POWER AMPLIFIERS



Model	Output Power R.M.S.	Dis- tortion Typical at 1KHz	Minimum Signal/ Noise Ratio	Power Supply Voltage	Size in mm	Weight in gms	Price + V.A.T.
HY30	15 W into 8 Ω	0.02%	80dB	-20 -0- +20	105x50x25	155	£6.34 + 95p
HY50	30 W into 8 Ω	0.02%	90dB	-25 -0- +25	105x50x25	155	£7.24 + £1.09
HY120	60 W into 8 Ω	0.01%	100dB	-35 -0- +35	114x50x85	575	£15.20 + £2.28
HY200	$\frac{120 \text{ W}}{\text{into 8 } \Omega}$	0.01%	100dB	-45 -0- +45	114×50×85	575	£18.44 + £2.77
HY400	$\begin{array}{c} 240 \ \mathbf{W} \\ \mathbf{into} \ 4 \ \boldsymbol{\Omega} \end{array}$	0.01%	100dB	-45 -0- +45	114×100×85	1.15Kg	£27.68 + £4.15

Load impedance - all models 4 - 16 A Input sensitivity - all models 500 mV Input impedance - all models 100KA Frequency response - all models 10Hz - 45Hz - 3dB



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R.C.S. LOUDSPEAKER BARGAINS 3 ohm. 6 × 41n. £1·50. 7 × 41n £1·50. 8 × 51n. £2·50. 6<u>1</u> £1·80. 81n, £2·60. 101n, £3. 121n. £4. 8 ohm. 2<u>1</u> n. £1·50. 51n. £1·50. 101n. £3. 121n. £4. 16 ohm. 6 × 41n. £1·50. 7 × 41n. £1·50. 51n. £1·50. 81n. £2·60. 101n. £3. 121n. £4. 10 × 61n. £3·50.

LOW VOLTAGE ELECTROLYTICS 1, 2, 4, 5, 8, 16, 25, 30, 50, 100, 200m F 15V 100, 500mF 12V 15p; 25V 20p; 50V 30p; 1200mF/76V 80p. 1000mF 12V 17p; 25V 35p; 50V 47p; 100V 70p; 2000mF 6V 25p; 25V 42p; 420mF/50V £1:30. 2500mF 50V 62p; 3000mF 25V 47p; 50V 65p. 3300mF 63V 21:20; 1700mF 63V £1:20; 2700mF/76V £1. 5000mF 6V 25p; 35V 85p. 5600mF/76V £1.57.

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16/350V		8+16/450V 50p	32+32/450V 75p						
32/500V		16+16/450V 50p	100+100/275V 85p						
50/350V	80p	32+32/350V 50p	150+200/275V 70p						
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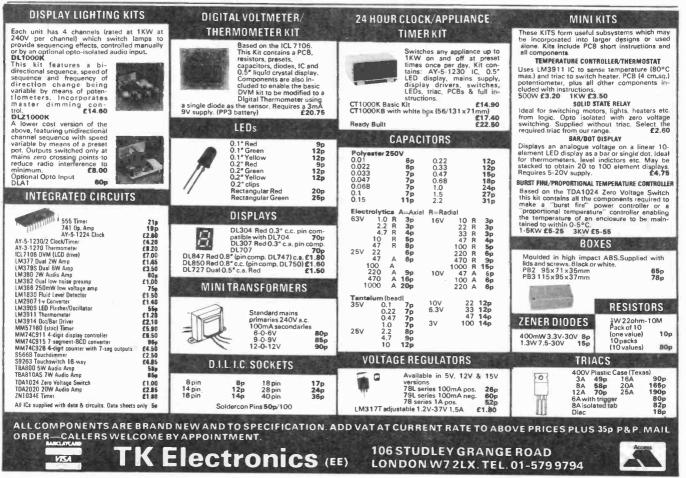
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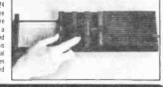


Everyday Electronics, March 1980



24 TUNE DOOR CHIMES

DOOR TUNES EIT 13 + VAT. Waddington's Videomaster announce a doorbell that doesn't Brringgg, Ding-Dong or Bzzzzz. Instead it plays 24 Herent classical and popular tunes. It will play the tune you select for your mond the season or the visitor you are expecting to call. Door tunes is not only great fun and a wonderful ice breaker, but is also very functionally and beautifully designed to enhance your home. There is something for the Queen. Door tunes is easy to install and has senarate controls for volume, tone and tempo



T.V. GAMES

PROGRAMMABLE £29.50 + VAT COLOUR CARTRIDGE T.V. GAME.

The TV game can be compared to an audio cassette deck and is programmed to play a multitude of different games in COLOUR, using various plug-in cartridges. At long last a TV game is available which will keep pace with improving nology by allowing you to extend your library of games the purchase of additional cartridges as new games tor with the pu are developed. Each cartridge contains up to ten different action games and the first cartridge containing ten sports games is included free with the console. Driver cartridges are currently available to enable you to play such games as Further cartridges are to be released later this year, including Tank Battle, Hunt the Sub and Target. The console comes complete with two removable joystick player controls to enable you to move in all four directions (upidown/right/left) and built into these joystick controls are ball serve and target fire buttons. Other features include several difficulty option switches, automatic on screen digital scoting and colour coding on scores and balls. Lifelike sounds are transmitted through the TV's speaker, Simulating the actual game being played. Manufactude by Waddington's Videomaster and 10 Game COLOUR SPORTSWORLD £22.50 + VAT. guaranteed for one year.



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CHESS COMPUTERS

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ng your own TV to display the board and pieces. Stat using your own IV to display the board and pieces. Star Chess is a new absorbing game for two players, which will interest and excite all ages. The unit plugs into the aerial socket of your IV set and displays the board and pieces in full colour for thack and whitel on your IV screen. Based on the moves of chess. It adds even more excitement and interest to the game. For those who have never played, Star Chess is a novel introduction to the classic game of chess. For the experienced chess player, there are whole new dimensions of unpredictability and chance added to the strategy of the game. Not only can pieces be taken in conventional chess type moves, but each piece can also exchange rocker fire with its opponents. The unit comes complete with a free 18V mains adaptor, full instructions and twelve months quarantee

CHESS CHALLENGER 7 - £85 65 + VAT. PLAY CHESS AGAINST THE COMPUTER.

The stylish, compact, portable console can be set to play at seven different levels of ability from beginner to expert including "Mate in two" and "Chess by mail". The computer Including "Mate in two" and "Lhess by mail". The computer will only make responses which obey international chess rules. Castling, on passant, and promoting a pawn are all included as part of the computer's programme. It is possible to enter any given problem from magazines or newspapers or alternatively establish your own board position and watch the computer react. The positions of all positions and watch the computer react. The positions of all passes on be unified by unified and the computer react. pieces can be verified by using the computor memory recall

Price includes unit with wood grained housing, and Staunton design chess pieces. Computer plays black or white and against itself and comes complete with a mains adaptor and 12 months guarantee.

OTHER CHESS COMPUTERS IN OUR RANGE INCLUDE: CHESS CHAMPION-6 LEVELS £47-39 + VAT. CHESS CHALLENGER - 10 LEVELS £138-70 + VAT. BORIS - MULTI-LEVEL TALKING DISPLAY £163-04 + VAT.

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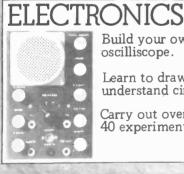
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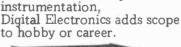
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A١

FM

FM radio control RX kit

- 8 Channel RC receiver (FM) Single IC RF/IF/Detector Single IC decoder 27MHz ceramic filter input FET RF stage with double tuned bandpass filter Dual ceramic filter IF Best quality SLM servo
 - connector block ONLY £16.10 inc VAT (kit) (includes new SLM case)

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211-24 Not illustrated here - but also now available is the DFM6. This is a vacuum fluorescent display -ersion of our immensely popular DFM3 (LCD), Resolution is 100Hz to 3.9999MHz, 39,999MHz, and 10kHz to 200.00MHz+; all standard IF offsets (inc. 10.7MHz on shortwave) are available via diode programming. UM1181 VHF band 2 VARICAP TUNERHEAD

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I circuit, with image/spurii better than -80dB, buffered LO MOSFET RF stage, FET IF preamp; tunes with only 1½ to 8v 3rd order intercept. 1off price £12,00 inc VAT. (100off/ OA) 5 tuned output,

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91072 AM RADIO TUNER MODULES · DC TUNED and DC SWITCHED Available February '80 All include buffered LO output, mechanical IF fitter (TOKO CFMQ) 1.10v tuning bias, switching by a single pole to earth A MW/LW (150 to 350kHz LW range) with farrite rod antenna B Az 'A' but also including SW1 or SW2 (specify.) SW1 = 1.8 to 4MHz SW2 = 5 to 10MHz C With both SW ranges Prices one off INC VAT 'A' £14.43 'B' £15.90 'C' £17.50 [Custom types OA]⁵,

There is a danger - when advertizing in some magazines - that because we do not find space to list everything we sell in every ad., that some readers forget about half the ranges we stock. So to summarize the general ranges: Chokes, coils for AM/FM/SW/ τοκο

MPX, Audio filters etc Filters: Ceramic for AM/FM, LC for FM, MPX etc. Polyvaricons ICs for radio, clock LSI, radio control, MPX decoders etc Dust iron cores for toroids Micrometals for resonant and EMI filters Toroid mounts Radio/audio/mpx linear ICs Hitachi 100W MOSFETs, small signal FETs, MOSFETs and bipolar

And the following groups of products from a broad range of sources:

Semiconductors -specializing in radio devices, Plessey SL1600, EUROPE's best selection of AM/FM and communications devices. Power MOSFETs, WORLD'S LOWEST NOISE AUDIO small signal transistors, BAR graph LED drivers for linear and log. CD4000 series CMOS, TTL/LPSNTTL, standard CD4000 series CMOS, TTL/LPSNTTL, standarc linears (741, 301, 3080 etc). MPUs, memories. Small signal transistors from AEG BC237/8/9 Yamilies etc. (1000 off BC239C : 5.2p ea) LEDs: AEG 3mm/5mm round, 2.5x5mm flat, red, greem, orange, yellow. The best prices you will find for quality products. <u>MOSFETs for RF signal processing</u>, including the BF960 UHF device, and 35K51 for VHF. Varicap diodes for 17:1 capacity ratio tuning

FREQUENCY READOUT LSI from OKI, with a one-chip answer to most digital frequency display needs (and various modules). Crystal and ceramic ladder filters from leading manufacturers, ferrite rods, various ferrite beads and a range of crystals for 'standard' frequencies and both AM and FM radio control at 27MHz.

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We supply parts for nearly all EE projects—for a detailed components list of this month's, and previous articles, please send SAE.



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V002 Twin type.2 meters 40 × 40mm and driver board, supplied with circuit and connexion data, £3:50. V003 New type, just in. Twin type moulded in one piece. 80 × 40mm (no driver board but suitable circuit supplied). £2:50.



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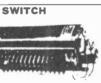
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DRILL CONTROLLER Electronically changes speed from approximately 10 revs to maximum. Full power at all speeds by finger-tip control. Kit includes all parts, case, everything and full instructions. 69.45 Made up model £2.00 extra

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service engineer. Price zs - ... TEST LEADS 5 pairs of different coloured leads. 10 leads in all, at each en is an insulated crocodile clip of the same colour as the lead 20 clips in all. This is invaluable for hook ups and will sav its cost in no time at all, Price £1 + 15p.

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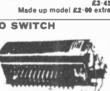
Price £1.78 + 39p. MAINS TRANSISTOR POWER PACK to operate transistor radio, cassettes, amplifiers, etc., take the place of any of the following batteries, PP1, PP3, PP4, PP6, PP9 and others. You can make voltage output anything from 3v to 16v at up to 300 mA. Complete kit but no case £1.75 + 28p. Case 75p + 13p.

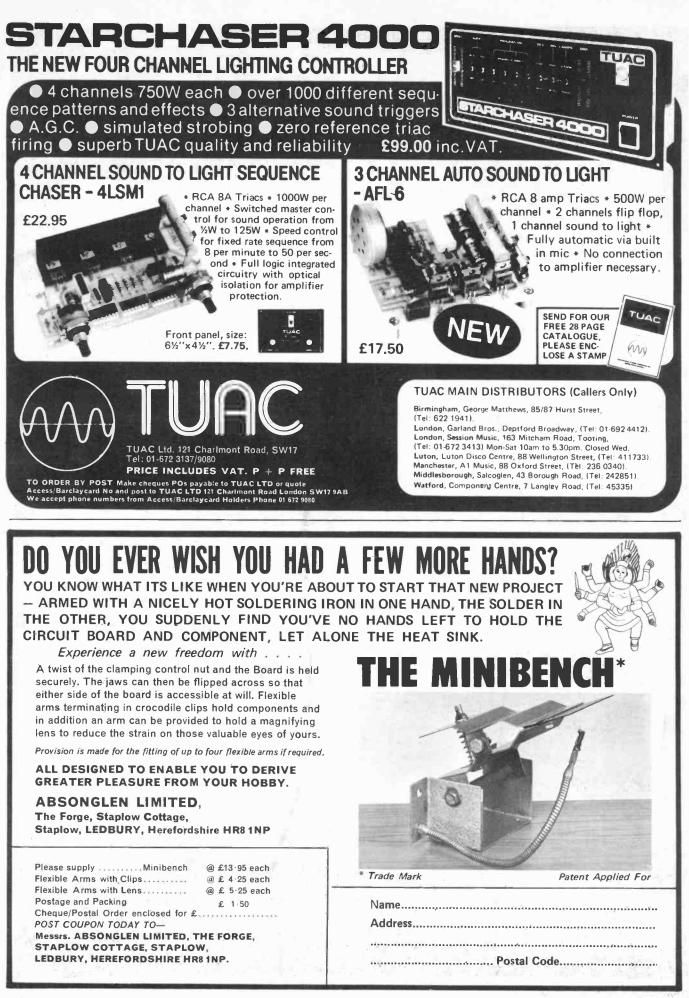
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with this kit. SkW BLOWER HEATER This kit consists of two tangential blowers, two 2.5 kW mineral filled heating elements, three level switch, thermal overload trip, mains input panel angle iron assembly frame and chassis. Outer case is not included but we do supply details of simple cases which you should be able to make. Price £11.58 + £1.75 + £2 post.

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Circuit diagram but no case. Price $\mathfrak{s}\mathfrak{L}$ + sup. FM TUNER Nicely boxed in wooden case and has a smart dial with cord drive and pointer, it covers the range 85-108 mhz., has an input sensitivity of 10uV, uses 6 transistors and 3 diodes. It is an excellent performer and when tested here we have had results quite equal to a tuner costing over £40. Price only £7:59 + £1.12 or with stereo decoder £11.59 + £1.72 + cost £1.58 £7·56 + £ cost £1·00.

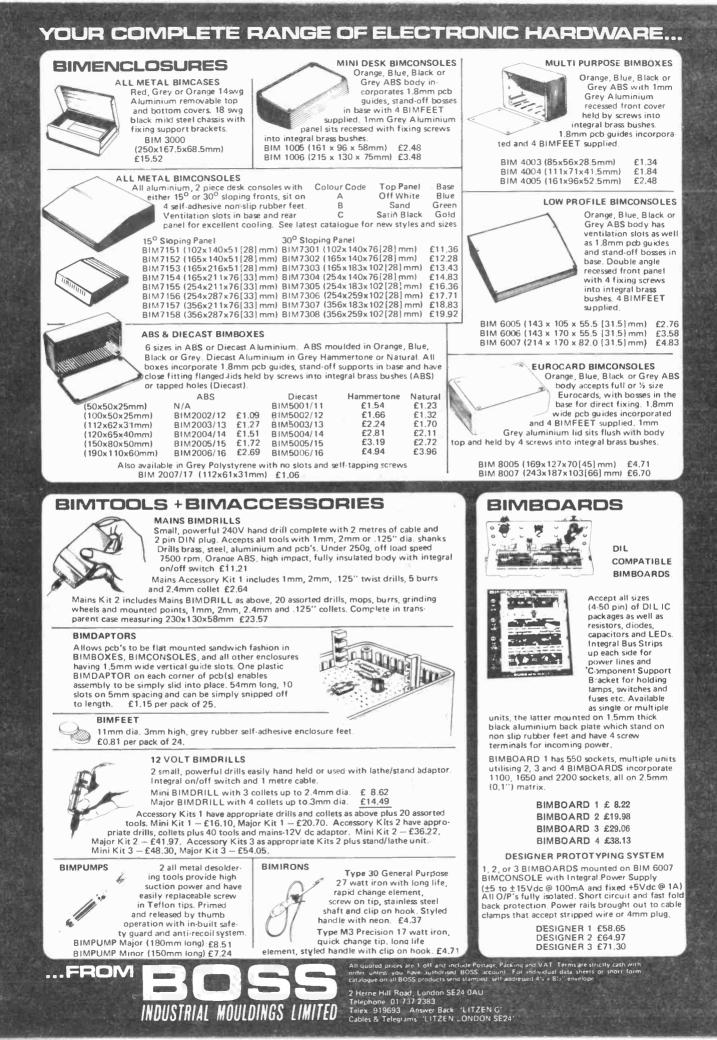






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8 pin 10p	20 pin	26p	C-15		380	a	21 ×	5	55 p
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C INCOM	0.00	-		W KIt	570		Note	ack cutting	105p
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.00r		50p	Stan		160		S.F.	Cutter	85p
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C126/7 C128 C176	20p 20p	CA30 CA30	46 80E	70p 72p	7400 7401		15p 15p	4017 4018	80p 80p
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D149 D161/2	70p 45p	CA30 CA31		375p 50p	7403 7404		15p 17p	4022 4023	1000
C107/8 C109	11p	ICL80	38	340p	7408		19p	4024	50p
C177/8	12p 17p	LF356 LM30		95p 36p	7410 7413		15p 30p	4027 4030	50p 55p
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CY71/2	22p	LM380)	900	7440		30p 17p	4069 4070	20p 22p
C131/2 IPSA12	50p 50p	LM381		160p 50p	7447A		60p	4081	22p
IP31A	58p	LM710		20p	7448 7450		80p 17p	4098 4411	120p
1P32A 1P33A	68p 90p	LM748	3	35p 70p	7470		360	4502	1200
1P34A	115p	LM390	9	70p 100p	7473		30p 24p	4503 4511	70p 150p
IP41A IP42A	65p 70p	LM391 LM413	11	130p	7475		36p 35p	4516	1100
TX108	12p	MC13	10P	120p 150p	7476 7483		90p	4520 4528	90p
N2219 A N2222 A	22p 20p	MC14 MC33	58 40 P	55p 120p	7486		34p	4584	90p
N2369 A	20p	M C33	60P	120p	7490 7489		33p 175p		
N2646 N2926	50 p 8 p	NE531 NE555		140p 22p	7492A 7493		46p 33p	VOLTAGE	OPE
N3053	25p	NE556	1	70p	74107		34p	1 Amp +ve	UR3
N3055 N3442	48p 140p	NE567 SN760		175p 175p	74121 74123		28p 55p	Plastic 5V	70 p
N3773	300p	SN760	13N	140n	74141		70p	8V	70 p
N3819 N3702/2	25p 12p	SN760	13ND	120p 120p	74154 74157		100p 70p	12V 15V	70p
V4123	27p	SN760	033N	175p	74160		100p	18V	90p
N4401 N5191	27p 83p	TBA6	41B11	225p 100p	74161 74164		100p 120p	24 V 1 Amp	900
N5194 N5245	90p 40p	TBA8 TCA9	10S 40E	110p 175p	74198		150p	5V 12V	90 p 90 p
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N5459 N6107 N6254	40p 130p	TL074 TL081		150p 48p	4000		MOS	OTHER	8
N6247	190p	TL084		130p 90p	4001 4009		19p 40p	LM317T LM323K	200p
N6290 0673	65p 75p	ZN414 ZN425		100p 400p	4010		50p	78H05	625p
0871/2	90 p	ZN103		200p	4011 4013		19p 50p	78MGT2C LM723	140p 37p
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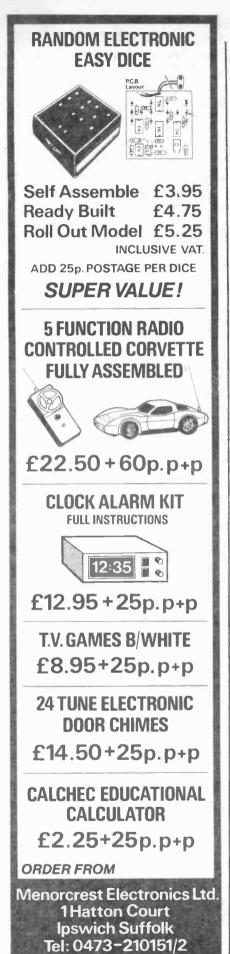
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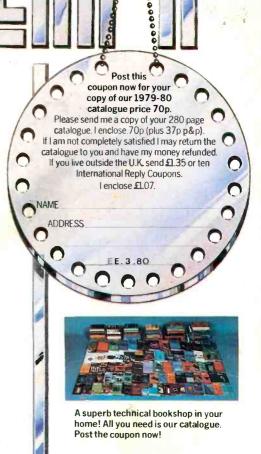
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