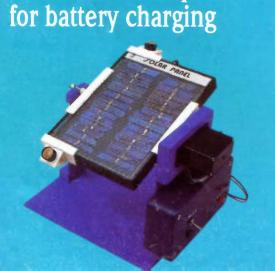
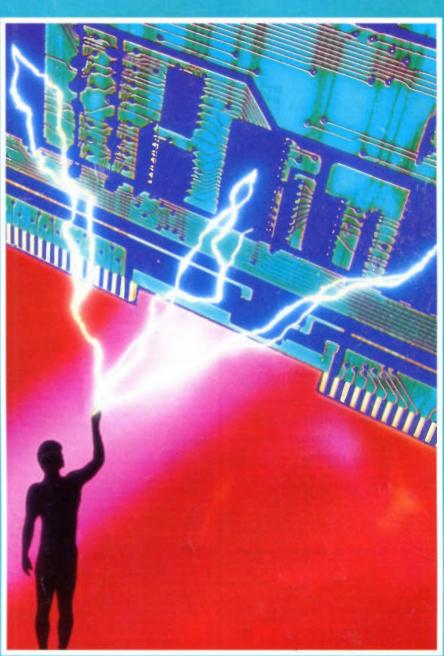
# With PRACTICAL E C P O C S

INCORPORATING ELECTRONICS MONTHLY

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# STATIC How to avoid the problems Personal Practice AMPLIFIER For budding rock stars **IICROCONTROLLED** 3-DIGIT TIMER Versatile up/down timer INFRA-RED REMOTE CONTROL Adds remote control to almost anything SOLAR SEEKER Motorised solar panel





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ENERGY BANK KIT 100 6"x6" 6v 100mA panels, 100 diodes, connection details etc. £69.95 ref EF112.

CCTV CAMERA MODULES 46X70X29mm, 30 grams, 12v 100mA. auto electronic shutter, 3.6mm F2 lens, CCIR, 512x492 pixels, video output is 1v p-p (75 ohm). Works directly into a scart or video input on a tv or video. IR sensitive. £79.95 ref EF137.

IR LAMP KIT Suitable for the above camera enables the camera to be used in total darkness! £5.99 ref EF138.

PASTEL ACCOUNTS SOFTWARE, does everything for all sizes of businesses, Includes wordprocessor, report writer, windowing, networkable up to 10 stations, multiple cash books etc. 200 page comprehensive manual, 90 days free technical support (0345-326009 try before you buy!) Current retail price is £129, ours? just £29 ref EF134. SAVE £100!!!

MINIMICRO FANS 12V 1.5' sq just £3.99 each. Ref EF199. CITOH PRINTERS 80 col, 9 pin matrix, serial/parallel, NLQ/draft,

3 mth warranty, good condition, £49 ref EF133.

MICROSOFT TRACKBALL AND MOUSE Combined unit with 4 buttons and trackball, PS2 type connector. Complete with storage bracket. Our price just £11.99 ref EF201.

REUSEABLE HEAT PACKS, Ideal for fishermen, outdoor enthusiasts elderly or infirm, warming food, drinks etc. defrosting pipes etc.reuseable up to 10 times, lasts for up to 8 hours per go, 2,000wh energy, gets up to 90 degC. Price is £12 ref EF129, rrp £37!

1.44MB3.6" DISC DRIVES Returns from a top PC manufactuer o they may need attention, bargain price £8.50 ea ref EF203.

1.2MB 5.26" DISC DRIVES Again returns somay need attention bargain price is £8.50 ref EF204. (1 of each 1.2+1.44£14.99 ref ef205 A4 DTP MONITORS Brand new, 300 DPI. Complete with diagram but no interface details (so you will have to work it out!) Bargain at just £7.99 each!!!! Ref EF186 OPD MONITORS 9' mono monitor, fully cased complete with rasterboard, switched mode psu etc. CGA/TTL Input (15way D), IEC mains. £15.99 ref DEC23. Price including kit to convert to composi monitor for CCTV use etc is £21.99 ref DEC24.

12V 2AMP LAPTOP psu's 110x55x40mm (includes standard IEC socket) and 2m lead with plug. 100-240v IP. £8.99 ref EF200. PC CONTROLLED 4 CHANNEL TIMER Control (or/off times etc) up to 4 items (8A 240v each) with this kit. Complete with Software, relays, PCB etc. £25.99 Ref 95/26

COMPLETE PC 300 WATT UPS SYSTEM Top of the range UPS system providing protection for your computer system and valuable software against mains power fluctuations and cuts. New and hoved. UK made Provides up to 5 mins running time in the event of complete power failure to allow you to run your system down correctly. SALE PRICE just £119.00.

RACAL MODEM BONANZA! 1 Racal MPS1223 1200/75 modern, telephone lead, mains lead, manual and comms software, the cheapest way onto the nett all this for just £13 ref DEC13.

HOW LOW ARE YOUR FLOPPIES? 3.5' (1.44) unbranded We have sold 100,000+ so ok! Pack of 50 £24,99 ref DEC16

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SHOP WOBBLERS!Small assemblies designed to take D size batteries and 'wobble' signs about in shops! £3.99 Ref SEP4P2.

RADIO PAGERSBrand new, UK made pocket pagers clearance price is just £4.99 each 100x40x15mm packed with bits! Ref SEP5. BULL TENS UNIT Fully built and tested TENS (Transcutaneous Electrical Nerve Stimulation) unit, complete with electrodes and full instructions. TENS is used for the relief of pain etc in up to 70% of sufferers. Drug free pain relief, safe and easy to use, can be used in conjunction with analgesics etc. £49 Ref TEN/1

COMPUTER RS232 TERMINALS. (LIBERTY)Excellent quality modern units. (like wyse 50, s) 2xRS232, 20 function keys, 50 thro to 38,400 baud, menu driven port, screen, cursor, and keyboard setup menus (18 menu's). £29 REF NOV4.

OMRON TEMPERATURE CONTROLLERS (E5C2).Brand new controllers, adjustable from 0 deg C to +100 deg C using graduated dial, 2% accuracy, thermocouple input, long life relay output .3A 240v o/o contacts. Perfect for exactly controlling a temperature, Normal trade £50+, ours £15. Ref E5C2.

ELECTRIC MOTOR BONANZA! 110x60mm.Brand new precision, cap start (or spin to start), virtually silent and features a moving outer case that acts as a fly wheel. Because of their unusual design we think that 2 of these in a tube with some homemade fan es could form the basis for a wind tunnel etc. Clearance price is just £4.99 FOR APAIR! (note-these will have to be wired in series for 240v operation Ref NOV1

MOTOR NO 2 BARGAIN 110x90mm. Similar to the above motor but more suitable for mounting vertically (ie turntable etc). Again you will have to wire 2 in series for 240v use. Bargain price is just £4.99 FOR A PAIR!! Ref NOV3.

## OMRON ELECTRONIC INTERVAL TIMERS. .....NEW LOW PRICES TO CLEAR!!!.....

Minature adjustable timers, 4 pole c/o output 3A 240v, HY1230S, 12vDC adjustable from 0-30 secs. £4.99 HY1210M, 12vDC adjustable from 0-10 mins. £4.99 HY1260M, 12vDC adjustable from 0-60 mins. £4.99 HY2460M, 24vAC adjustable from 0-60 mins. £2.99 HY241S, 24vAC adjustable from 0-1 secs. £2.99 HY2460S, 24vAC adjustable from 0-60 secs, £2.99 HY243H, 24vAC adjustable from 0-3 hours. £2.99 HY2401S, 240v adjustable from 0-1 secs. £4.99 HY2405S, 240v adjustable from 0-5 secs. £4.99 HY24060m, 240v adjustable from 0-60 mins. £6.99 PC PAL VGA TO TV CONVERTER Converts a colour TV into a basic VGA screen. Complete with built in psu, lead and s/ware. £49.95, ideal for laptops or a cheap upgrade. We also can supply this in kit form for home assembly at £34.95 ref EF54.

DRINKING BIRD Remember these? hook onto wine glass (supplied) and they drink, standup drink, standup ETC! £4 each Ref EF1 EMERGENCY LIGHTING UNIT Complete unit with 2 double bulb floodlights, built in charger and auto switch. Fully cased. 6v 8AH lead acid regid (secondhand) £4 ref MAG4P11

GUIDED MISSILE WIRE. 4 200 metre reel of ultra thin 4 core nsulated cable, 28lbs breaking strain, less than 1mm thick! Ideal alarms, intercoms, fishing, dolls house's etc. £14.99 ref MAG15P5 ASTEC SWITCHED MODE PSU BM41012 Gives +5 @ 3.75A +12@1.5A.-12@.4A. 230/110, cased, BM41012, £5,99 ref AUG6P3. AUTO SUNCHARGER 155x300mm solar panel with diode and 3 metre lead fitted with a cigar plug. 12v 2watt. £9.99 ea ref AUG10P3. FLOPPY DISCS DSDD Top quality 5.25" discs, these have been written to once and are unused. Pack of 20 is £4 ref AUG4P1.

ECLATRON FLASH TUBE As used in police car flashing lights etc, full spec supplied, 60-100 flashes a min. £9.99 ref APR10P5 24v AC 96WATT Cased power supply. New. £13.99 ref APR14.

MILITARY SPECGEIGER COUNTERS Unused anstraightfrom

OUTDOOR SOLAR PATH LIGHT Captures sunlight during the day and automatically switches on a built in lamp at dusk. Complete with sealed lead acid battery etc.£19,99 ref MAR20P1.

ALARM VERSION Of above unit comes with built in alarm and pir to deter intruders. Good value at just £24,99 ref MAR25P4

CARETAKER VOLUMETRIC Alarm, will cover the whole of the ground floor against forcred entry. Includes mains power supply and integral battery backup. Powerful internal sounder, will take external bell if req'd. Retail £150+, ours? £49.99 ref MAR50P1.

TELEPHONE CABLE White 6 core 100m reel complete with a pack of 100 clips, Ideal phone extris etc. £7.99 ref MAR8P3

MICRODRIVE STRIPPERS Small cased tape drives ideal for stripping, lots of useful goodies including a smart case, and lots of components. £2 each ref JUN2P3. Box of 10 just £9.99 ref EF207. SOLAR POWER LAB SPECIAL You get TWO 6'x6" 6v 130mA solar cells, 4 LED's, wire, buzzer, switch plus 1 relay or motor. Superb value kit just £5.99 REF: MAG6P8

BUGGING TAPE RECORDER Small voice activated recorder micro cassette complete with headphones, £28.99 ref MAR 29P1 ULTRAMINIBUG MIC 6mmx3.5mm made by AKG, 5-12v electret condenser. Cost £12 ea, Ours? just four for £9.99 REF MAG10P2. RGB/CGA/EGA/TTL COLOUR MONITORS 12' in good on. Back anodised metal case. £79 each REF JUN79

ANSWER PHONES Returns with 2 faults, we give you the bits for 1 fault, you have to find the other yourself. BT Response 200's £18 ea REF MAG18P1, PSU£5 ref MAG5P12.

SWITCHED MODE PSU ex equip, 60w +5v @5A, -5v@.5A, +12v@2A,-12v@.5A 120/220v cased 245x88x55mm IECinput socket £6.99 REF MAG7P1

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ACORN ARCHIMEDES PSU +5v @ 4.4A. on/off sw uncased, electable mains input 145x100x45mm £7 REF MAG7P2 9v DC POWER SUPPLY Standard plug in type 150ma 9v DC with

lead and DC power plug, price for two is £2,99 ref AUG3P4 AA NICAD PACK encapsulated pack of 8 AA nicad batteries (tagged) ex equip, 55x32x32mm, £3 a pack, REF MAG3P11

13.8V 1.9A psu cased with leads. Just £9.99 REF MAG10P3 PPC MODEM CARDS. These are high spec plug in cards made for the Amstrad laptop computers, 2400 baud dial up unit complete with leads. Clearance price is £5 REF: MAG5P1

INFRA RED REMOTE CONTROLLERS Originally made for hi spec satellite equipment but perfect for all sorts of remote control projects. Our dearance price is just £2 REF: MAG2

200 WATT INVERTER Converts 10-15v DC into either 110v or 240v AC. Fully cased 115x36x156mm, complete with heavy duty power lead, cigar plug, AC outlet socket Auto overload shutdown, auto short circuit shut down, auto input over voltage shutdown, auto input under voltage shut down (with audible alarm), auto temp control, unit shuts down if overheated and sounds audible alarm. Fused reversed polarity protected, output frequency within 2%, voltage within 10%. A well built unitatan keen price Just £64.99 ref AUG65. UNIVERSAL SPEED CONTROLLER KIT Designed by us for

the C5 motor but ok for any 12v motor up to 30A. Complete with PCB etc. A heat sink may be required. £17.00 REF: MAG17 MAINSCABLE Precut black 2 core 2 metre lengths ideal for repairs, projects etc. 50 metres for £1.99 ref AUG2P7.

COMPUTER COMMUNICATIONS PACK Kit contains 100m of 6 core cable, 100 cable clips, 2 line drivers with RS232 interfaces and all connectors etc. Ideal low cost method of communicating between PC's over a long distance. Complete kit £8.99.

ELECTRIC MOTOR KIT Comprehensive educational kit ins all you need to build an electric motor. £9.99 ref MAR10P4. VIEW DATA SYSTEMS made by Phillips, complete with internal 1200/75 modem, keyboard, psu etc RGB and composite outputs, menu driven, autodialler etc. £18 each Ref EF88.

BOOMERANG High tech, patented poly propylene, 34cm wing pan. Get out and get some exercise for £4.99 ref EF83

AIR RIFLES .22 As used by the Chinese army for training puposes g there is a lot about! £39.95 Ref EF78. 500 pellets £4.50 ref EF80. PLUG IN POWER SUPPLYS Plugs in to 13A socket with output lead, three types available, 9vdc 150mA £2 ref EF58, 9vdc 200mA £2.50 ref EF59, 6.5vdc 500mA £3 ref EF61,

VIDEO SENDER UNIT. Transmits both audio and video signals from either a video camera, video recorder, TV or Computer etc to any standard TV set in a 100' rangel (tune TV to a spare channel) 12v DO BOME OF OUR PRODUCTS MAY BE UNLICENSABLE IN THE U

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\*FM CORDLESS MICROPHONE Small hand held unit with a

500' range! 2 transmit power levels. Reqs PP3 9v b to any FM receiver. Price is £15 REF; MAG15P1 9v battery. Tuneable

LOW COST WALKIE TALKIES Pair of battery operated units with a range of about 200', ideal for garden use or as an educational toy. Price is £8 a pair REF: MAG 8P1 2 x PP3 reg'd.

MINATURE RADIO TRANSCEIVERS A pair of walkie talkies with a range up to 2 km in open country. Units measure 22x52x155mm cases and earp'ces. 2xPP3 req'd. £30.00 pr.REF: MAG30 COMPOSITE VIDEO KIT. Converts composite video into separate H sync, V sync, and video. 12v DC. £8.00 REF: MAG8P2.

LQ3600 PRINTER ASSEMBLIES Made by Amstrad they are entire mechanical printer assemblies including printhead, stepper motors etc etc in fact everything barthe case and electronics, a good stripper! £5 REF: MAGSP3 or 2 for £8 REF: MAG8P3

LED PACK of 100 standard red 5m leds £5 REF MAG5P4

UNIVERSAL PC POWER SUPPLY complete with flyleads, switch, fan etc.200w at £20 REF; MAG20P3 (265x155x125mm). GYROSCOP EAbout 3° high and an excellent educational toy for all

ages! Price with instruction booklet £6 Ref EF15. FUTURE PC POWER SUPPLIES These are 295x135x60mm, 4 drive connectors 1 mother board connector, 150watt, 12v fan, iec

inlet and on/off switch. £12 Ref EF6. VENUS FLYTRAP KIT Grow your own carnivorous plant with this

PC POWER SUPPLIES (returns) These are 140x150x90mm, o/ ps are +12,-12,+5 and -5v. Built in 12v fan. These are returns so they may well need repairing! £3.50 each ref EF42.

FM TRANSMITTER KIT housed in a standard working 13A adapter!! the bug runs directly off the mains so lasts forever why pay £700? or price is £15 REF: EF62 Transmits to any FM radio. (this is with full instructions.)

\*FM BUG KIT New design with PCB embedded coil for extra stability. Works to any FM radio. 9v battery req'd. £5 REF: MAG5P5 \*FM BUG BUILT AND TESTED superior design to kit, Supplied to detective agencies, 9v battery req'd, £14 REF; MAG14

TALKING COINBOX STRIPPER onginally made to retail at £79 each, these units are designed to convert an ordinary phone into a payphone. The units have the locks missing and sometimes broken hinges. However they can be adapted for their original use or used for something else?? Price is just £3 REF: MAG3P1

TOP QUALITY SPEAKERS Made for HI Fi televisions these are 10 watt 4R Jap made 4' round with large shielded magnets. Good quality. £2 each REF: MAG2P4 or 4 for £6 REF: MAG6P2

TWEETERS 2" diameter good quality tweeter 140R (ok with the eaker) 2 for £2 REF: MAG2P5 or 4 for £3 REF: MAG3P4 AT KEYBOARDS Made by Apricot these quality keyboards need just a small mod to run on any AT, they work perfectly but you will have to put up with 1 or 2 foreign keycaps! Price £6 REF: MAG6P3

DOS PACKS Microsoft version 3.3 or higher complete with all manuals or price just £5 REF: MAG5P8 Worth it just for the very comprehensive manual! 5.25" only.

GAS HOBS Brand new made by Optimus, basic three burner suitable for small flat etc bargain price just £29.95 ref EF73.

GAT AIR PISTOL PACK Complete with pistol, darts and pellets EF82 extra pellets (500) £4.50 ref EF80.

DOS PACK Microsoft version 6 with manual £9,99 3.5" ref EF209 WINDOWS 3.1 3.5" with manual £24.99 ref EF210.

NOVELL NTEWARE LITE (network s/ware) £24,99 ref EF211. PIR DETECTOR Made by famous UK alarm manufacturer these are hi spec, long range internal units, 12v operation. Slight marks on case and unboxed (although brand new) £8 REF: MAG8P5

MOBILE CARPHONE £5.99 Well almost! complete in carphone excluding the box of electronics normally hidden under seat. Can be made to illuminate with 12v also has built in light sensors odisplay only illuminates when dark. Totally convincing! REF: MAG6P6 6"X12" AMORPHOUS SOLAR PANEL 12v 155x310mm

30mA. Bargain price just £5.99 ea REF MAG6P12.

FIRRE OPTIC CARLE RUMPER PACK 10 metres for £4.99. ref MAG5P13 ideal for experimentersl 30 m for £12.99 ref MAG13P1 HEATSINKS (finned) TO220, designed to mount vertically on a pcb 50x40x25mm you can have a pack of 4 for £1 ref JUN1P11.

STROBE LIGHT KIT Adjustable from 1 hz right up to 60hz! ctronic asssembly kit with full Instructions) £16 ref EF28 ROCK LIGHTS Unusual things these, two pieces of rock that glow

when rubbed together! belived to cause rain!£3 a pair Ref EF29. AMSTRAD GX4000 games machines, returns, untested, sold as

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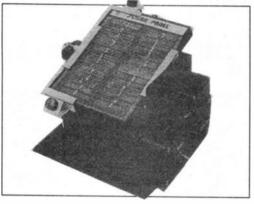
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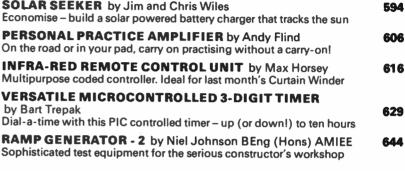
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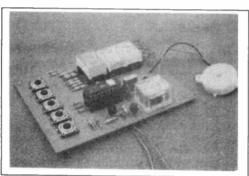
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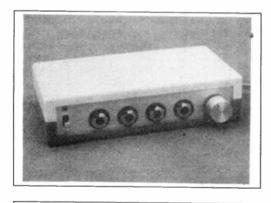
**EVERYDAY** 





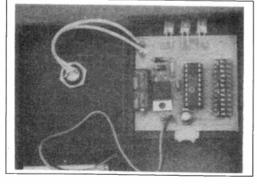
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Our September '95 Issue will be published on Friday, 4th August 1995. See page 583 for details.

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# One of the highest specification monitors you will ever see 🔈 At this price - Don't miss it!! 🙍



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PHILIPS HCS35 (same style as CM8833) attractively styled 14" colour monitor with both RGB and standard composite 15.825 Khz video Inputs via SCART socket and separate phono jacks. Integral audio power amp and speaker for all audio visual uses. Will connect direct to Amige and Atart BBC computers, ideal for most colour cameras. High quality with many features such as front concealed flap controls, VCR correction button etc. Good used condition - fully tested with a 90 day guarantee.

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Special Offer save £16.95 - Order TELEBOX ST & HCS35 together - giving you a quality colour TV & AV system for Only £122.50 (E)

KME 10" high definition colour monitors. Nice tight 0.28" dot pitch for superb clarity and modem styling. Operates from any 15.625 khz sync RGB video source, with RGB analog and composite sync such as Alari, Commodore Amiga, Acorn Archimedes & BBC. Measures only 13%" x 12" x 11". Only £125 (E) Good used condition. 90 day guarantee.

KME 10" as above for PC EGA standard £145.00 (E)

PHILIPS 10-38 above for PC EGA standard 2:145.00 (c)

PHILIPS 14-531 Ultra compact 6" colour video monitor with standard composite 15.625 Khz video input via SCART socket. Idea for all monitoring / security applications. High quality, ex-equipment fully tested with a 90 day guarantee (possible minor screen burns) in attractive square black plastic case measuring W10" x H10" x 13% D. Mains powered Limited Quantity - Only £79.00 (D)

# 20" 22" and 26" AV SPECIALS

Superbly made UK manufacture. PIL all solid state colour monitors, complete with composite video & optional sound inputs. Attractive teak style case. Perfect for Schools, Shops, Disco, Clubs, etc.in EXCELLENT little used condition with full 90 day guarantee.

20"....£135 22"....£155 26"....£185(F)

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Virtuelly every type of power supply you cen imegine.Over 10,000 Power Supplies Ex Stock Cell for info / liet.

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Less than Half Price!

Top quality 19° rack cabinets made in UK by Optima Enclosures Ltd. Units feature designer, smoked acrylic lockable front door, full height lockable half louvered back door and removable side panels. Fully adjustable internal fixing struts, ready punched for any configuration of equipment mounting plus ready mounted integral 12 way 13 amp socket switched mains distribution strip make these racks some of the most verastile we have ever sold. Racks may be stacked side by side and therefore require only two side panels to stand singly or in bays

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Over 1000 racks in all sizes 19" 22" & 24" 3 to 44 U. Available from stock !! Call with your requirements.

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TOUCH SCREEN SYSTEM
The ultimate in Touch Screen Technology made by the experts MicroTouch - but sold at a price below cost It System consists of a flat translucent glass laminated panel measuring 29.5 x 23.5 cm connected to a PCB with on board sophisticated electronics. From the board comes a standard serial R9232 or TTL output. The output continuously gives simple serial data containing positional X & Y co-ordinates as to where a finger is touching the panel - as the finger moves, the data instantity changes. The X & Y information I signed at an incredible metrix resolution of 1024 x 1024 positions over the screen size Iff So, no position, however small fails detection. A host of available translation software enables direct connection to a PC for a myriad of applications including: control panels, pointing devices, POS systems, controllers for the disabled or computer un-trained etc etc Imagine using your finger in 'Windows' instead of a mouse II (a driver is indeed available I) The applications for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations III and the product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for this emaxing product are only limited by your Imaginations for the first product and the product are only limited by your Imaginations for the product are only limited by your Imaginations for the product are only limited by your Imaginations for the product are only limited

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NTEL 'ABOVE' Memory Expansion Board. Full length PC-XT and PC-AT compatible card with 2 Mbytes of memory on board.
Card is fully selectable for Expanded or Extended (286 processor and above) memory. Full data and driver disk supplied. In good used condition fully tested and guaranteed.
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Give all-round protection to your garden shed, garden furniture, garage and home. A mains powered intruder alarm with battery back-up. Features a hard-to-bypass balanced loop sensor system which interfaces to alarm bells, sirens and P.I.R. detectors.

# **VANDATA**

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# SIMPLE THEREMIN

Just wave your hand in the air to make music. The Theremin saw a resurgence of interest following last year's Channel 4 documentary on the Russian Leon Theremin who invented the first successful electronic musical instrument back in the 1920s. Next month we present a simple battery powered version for everyone's enjoyment. The fascination of the Theremin is that it is played without physical contact.



# LOW RANGE OHMMETER ADAPTOR

Ever tried measuring very low value resistances using a standard digital multimeter? This simple, inexpensive adaptor allows a DMM to measure resistance values of less than 1 $\Omega$  with two decimal place indication. The unit has two ranges, 0 to 1 $\Omega$  and 0 to 10 $\Omega$ . It is easy to build and calibrate.

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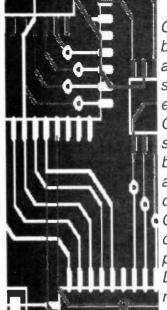
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SEPTEMBER ISSUE ON SALE FRIDAY, 4th AUGUST

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Electronics World & Wireless World Jan 1995

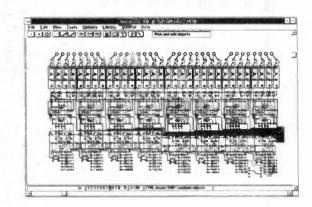
PRO+ now with SPICE file export & SpiceAge DDE link!



Quickroute 3.0 for Windows 3.1 has been designed from the start to be as easy to use as possible, without sacrificing the power professional engineers need to get the job done. Quickroute is available with schematic capture, support for busses & power rails, 1-8 layer auto-routing, SPICE file export, and our new extended library pack. Quickroute 3.0 PRO+ can also connect to the Windows simulation package SpiceAge using Windows DDE. Contact POWERware for more details!

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**Integrated Schematic and PCB Design for Windows 3.1** 





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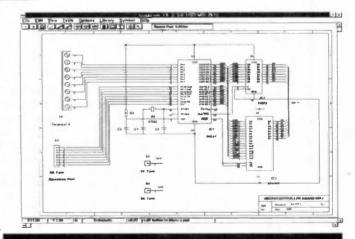
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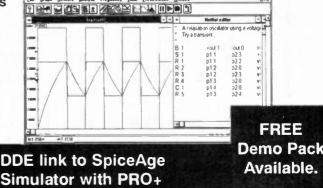
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# ELECTRONICS SOFTWARE

If you are looking for a means of improving your knowledge of the basics of electronics then this software is for you.

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Now an extended range of electronics topics with an improved graphics presentation, available to electronics hobbyists, students, schools, colleges, and used for training within industry throughout the UK and overseas. A fully interactive electronics program where the user selects from over 250 analogue and digital topics. Circuit diagrams, wave forms, phase angles, voltages and currents or logic states are shown, drawn to scale, in full colour. Formulae are given demonstrating all the calculation steps, exactly as in a textbook, using your input values.

An 'on screen' electronics package including circuit theory

An 'on screen' electronics package including circuit theory to enable a learning through doing approach to encourage experimentation. For the young student, mature hobbyist or the engineer that just needs to keep up-to-date in an easy and

enjoyable way.

Having reviewed a dozen, or more, educational software packages designed to "teach" electronics, I was more than a little sceptical when I first heard about Electronics Principles: there seemed to be little that could be done that has not been done elsewhere. When I started to use the package my views changed. Indeed, I was so impressed with it that I quickly came to the conclusion that readers should have an opportunity to try the package out for themselves! — MIKE TOOLEY B.A. Dean of Faculty of Technology, Brooklands Technical College.

# Complete package Still only £49.95

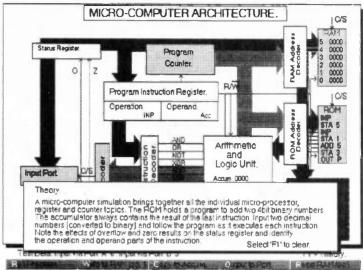
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thus encouraging experimentation in circuit design.

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# GCSE ELECTRONICS VERSION 2.0

Designed to complement the EPE Teach-In No. 7 publication, which contains a complete electronics course aimed at GCSE and A Level students, this software can also be used as a stand alone learning package for everyone interested in electronics. From simple d.c. current flow and Ohm's Law it covers the whole GCSE syllabus including a.c., semiconductors, op.amps, digital electronics, radio and fibre optic communications, digital numbering systems and goes on to look at microprocessor c.p.u.s, 6502 addressing modes and instruction sets. It even produces a program listing, which can be saved or retrieved from hard disk, when any of the instructions from the listed instruction set are input, thus allowing the user to monitor the registers for each of the addressing modes.

The original version is now widely used in schools and colleges throughout the UK.

An interactive, user friendly medium which assists learning in an enjoyable and interesting way.

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A series of programs covering all the major topics required by the school syllabus. Designed to be user friendly enabling you to study or revise in what we believe is an interesting and enjoyable way. There are nearly one hundred and fifty menu driven screens with interactive graphics, enabling a "learning through doing" approach to encourage experimentation. Now being used in many schools throughout the UK.

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These programs require a PC (or fully compatible system) running DOS or Windows with an 80286 or better processor and VGA (ideally colour) graphics. In addition you must have 4Mb of hard disk space, a high density (1.44Mb) floppy drive and at least 640K of RAM. We also recommend the use of a mouse. Site licences are available – please enquire.

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Add £2 per order for UK post and packing. Make cheques payable to EPT Educational Software.

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# SURVEINANCE PROFESSIONAL QUALITY KITS

Whether your requirement for surveillance equipment is amateur, professional or you are just fascinated by this unique area of electronics SUMA DESIGNS has a kit to fit the bill. We have been designing electronic surveillance equipment for over 12 years and you can be sure that all our kits are very well tried, tested and proven and come complete with full instructions, circuit diagrams, assembly details and all high quality components including fibreglass PCB. Unless otherwise stated all transmitters are tuneable and can be received on an ordinary VHF FM radio.

# Genuine SUMA kits available only direct from Suma Designs. Beware inferior imitations!

## **UTX Ultra-miniature Room Transmitter** Smallest room transmitter kit in the world! Incredible 10mm x 20mm including mic. 3-12V operation. 500m range. £16.45 MTX Micro-miniature Room Transmitter Best-selling micro-miniature Room Transmitter

# Just 17mm x 17mm including mic. 3-12V operation. 1000m range.....£13.45 STX High-performance Room Transmitter

Hi performance transmitter with a buffered output stage for greater stability and range. Measures 22mm x 22mm including mic. 6-12V operation, 1500m range .....

## **VT500 High-power Room Transmitter**

Powerful 250mW output providing excellent range and performance. Size 20mm x 40mm. 9-12V operation. 3000m range......

# **VXT Voice Activated Transmitter**

Triggers only when sounds are detected. Very low standby current. Variable sensitivity and delay with LED indicator. Size 20mm x 67mm. 9V operation. 1000m range...£19.45

## **HVX400 Mains Powered Room Transmitter**

Connects directly to 240V AC supply for long-term monitoring. Size 30mm x 35mm. 500m range.

# **SCRX Subcarrier Scrambled Room Transmitter**

Scrambled output from this transmitter cannot be monitored without the SCDM decoder connected to the receiver. Size 20mm x 67mm. 9V operation, 1000m range......£22.95

# **SCLX Subcarrier Telephone Transmitter**

Connects to telephone line anywhere, requires no batteries. Output scrambled so regulres SCDM connected to receiver. Size 32mm x 37mm. 1000m range ......£23.95

## **SCOM Subcarrier Decoder Unit for SCRX**

Connects to receiver earphone socket and provides decoded audio output to headphones. Size 32mm x 70mm. 9-12V operation......

# ATR2 Micro Size Telephone Recording Interface

Connects between telephone line (anywhere) and cassette recorder. Switches tape automatically as phone is used. All conversations recorded. Size 16mm x 32mm. Powered from line



# DLTX/DLRX Radio Control Switch

Remote control anything around your home or garden, outside lights, alarms, paging system etc. System consists of a small VHF transmitter with digital encoder and receiver unit with decoder and relay output, momentary or afternate, 8-way dil switches on both boards set your own unique security code. TX size 45mm x 45mm. RX size 35mm x 90mm. Both 9V operation, Range up to 200m.

Complete System (2 kits) Individual Transmitter DLTX. Individual Receiver DLRX

## MBX-1 Hi-Fi Micro Breadcaster

Not technically a surveillance device but a great ideal Connects to the headphone output of your Hi-Fi, tape or CD and transmits Hi-Fi quality to a nearby radio. Listen to your favourite music anywhere around the house, garden, in the bath or in the garage and you don't have to put up with the DJ's choice and boring waffle. Size 27mm x 60mm. 9V operation. 250m range .....

## **UTLX Ultra-miniature Telephone Transmitter**

Smallest telephone transmitter kit available. Incredible size of 10mm x 20mm! Connects to line (anywhere) and switches on and off with phone use. All conversation transmitted. Powered from line. 500m range..

# TLX700 Micro-miniature Telephone Transmitter

Best-selling telephone transmitter. Being 20mm x 20mm it is easier to assemble than UTLX. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted. Powered from line, 1000m range .....

## STLX High-performance Telephone Transmitter

High performance transmitter with buffered output stage providing excellent stability and performance. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted. Powered from line. Size 22mm x 22mm.

## TKX900 Signalling/Tracking Transmitter

Transmits a continous stream of audio pulses with variable tone and rate. Ideal for signalling or tracking purposes. High power output giving range up to 3000m. Size 25mm x 63mm, 9V operation...

CD400 Pocket Bug Detector/Locator
LED and piezo bleeper pulse slowly, rate of pulse and pitch of tome increase as you approach signal. Gain control allows pinpointing of source, Size 45mm x 54mm. 9V operation

# **CD600 Prefessional Bug Detector/Locator**

Multicolour readout of signal strength with variable rate bleeper and variable sensitivity used to detect and locate hidden transmitters. Switch to AUDIO CONFORM mode to distinguish between localised bug transmission and normal legitimate signals such as pagers, cellular, taxis etc. Size 70mm x 100mm. 9V operation ....

# QTX180 Crystal Controlled Room Transmitter

Narrow band FM transmitter for the ultimate in privacy. Operates on 180 MHz and requires the use of a scanner receiver or our QRX180 klt (see catalogue). Size \$40.95 20mm x 67mm, 9V operation, 1000m range...

# QLX180 Crystal Controlled Telephone Transmitter

As per QTX180 but connects to telephone line to monitor both sides of conversattions. 20mm x 67mm. 9V operation. 1000m range.....

# **QSX180 Line Powered Crystal Controlled Phone Transmitter**

As per QLX180 but draws power requirements from line. No batteries required. Size 32mm x 37mm. Range 500m...

# **QRX180 Crystal Controlled FM Receiver**

For monitoring any of the 'Q' range transmitters. High sensitivity unit. All RF section supplied as a pre-built and aligned module ready to connect on board so no difficulty setting up. Outpt to headphones, 60mm x 75mm, 9V operation.....

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# FANS







X6085

X6082 120x120x38mm 220V ac. various makes but all diecast frame and metal blades SALE £2.00

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X6086 170mm dia x 51mm deep, but with flattened sides 150mm across, this fan is made by Comair Rotron, Rated 24V DC 1A **SALE £3.50** 

# **POWER SUPPLIES**



MACINTOSH

X9012 Shiva EtherPort SF This brand new

boxed set of parts from Novell for networking Apple Macintosh computers is

fantastic value. You get the emerport ac Controller Board, the Etherport SE Access Board, the software on a 3.5" disk, an Installation Guide and a User's Guide. Wonderwhat the original cost was? Our

normal price £14.95

You get the Etherport SE

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professional fist microphone

Superb quality and value with

SALE PRICE £19.95

EXCELLENT

with push to talk switch

MICROPHONE

**25961** Shure 5278

all these features:

List Price is an A. Dynamic 8. Triple flex heavy duty cord

front panel. Comes in flat pack for self

assembly. Cat price £10.99

X5729 Set of 3 back to back connectors: 25W D socket - 25W D socket 25W D socket - 25W D plug 25W D socket - 35W Centronics plug

These are not 'straight through' connectors, but covers are easily removable to wire in any way desired.

SALE PRICE

£2.95

£71.67! C. Noise cancelling

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Our Price: £29.95

29157 One of the best power supplies we've seen for the money - this 397 watt switch mode beauty is one of the highest quality, made by Delta Bectronics Inc. Removed from Ingriest quarty, made by Deta Bectronics Inc. Removed from equipment, but in excellent condition (less than a year old!) the unit is totally enclosed in a steel case 340x152x152mm. It has an EC mains inlet with suppressor fitted and on/off mains rocker switch, and all outputs are on leads with power connectors. Now for the spec:

Inputs: 100-120V @ 10A or 200-240V @ 6A, Outputs
+5V @ 40A; +12V @ 15A; -5V @ 1A; -12V @ 1A
switchable on front panel. A 12Vdc 120x120mm fan is
fitted at the rear of the case. Current distributor price of a
unit of this ilk would be around £400.

# ETHERPORT FOR NICAD BATTERIES



Z4150 Back in stock! popular ex-radio batteries containing 8xAA NiCads. Generally NiCaus.
rejected because physical damage but may be the there may be the occasional duff cell. Total capacity. 10V @ 500mA. Price £1.60

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X6016 Brand new, boxed SMPSU by ASTEC. Mains in, -12V 0.2A +12V 2A and +5V 5A out 44W max PCB 160x100mm, £9.95 44W max PCB 160x100mm.

# **BACK IN STOCK**



Z5789 We've been very lucky to secure a further fev hundred of this extremely popular and useful gear set - and at the same price, too! There are 7 items.(a) Steel 56.6mm dia 3,7mm thick with boss 13,4mm dia and 7,7mm thick 5mm hole (b) Two of these: steel, 31.5mm dia, 6mm thick with boss 13,4mm dia 9.4mm thick 6mm hole.(c) & (d) Steel rack 86mm long and pinlon with 9mm hole(e) & (f) Plastic bevel gears giving right angle drive. The larger is 36.4mm dia with steel boss 9mm dia and 5mm hole, and the smaller is 18.5mm dia with 2mm hole.As can be si from the pic, they can all be put together in a chain if

# Case offer **VERSATILE** £3.00 CASE F660D Steel case 230x190x56mm with all

INSTRUMENT £2.95

X6147 Versatile soft padded instrument case measuring 250x250x80mm, with expanding side pocket 175x80x90mm. Both have veloro fasteners. On one side are two apertures, one 15x15mm and the other 80x40mm covered by a leatherette flap. Strong carrying handle, Grey material with black edging. Stacks of uses - toolbag, electrical/electronic equipment, sandwiches/drinks, knitting, cassettes (holds 36), CDs (14), school books

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60 watt mains transformer, suitable for PCB mounting or bolted to a chassis. Two main windings: 12V 2A and 16V 2A. There are also two low current windings - 5V 50mA and 38V 50mA 50mA and Overall size 64x53x65mm.
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Z5874 Great value on this

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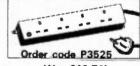


Z3100 Car VFD. This shows the plan view of a saloon and indicates whether doors are shut, lights are on etc. With data sheet £2.00



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# 4 WAY TRAILING SOCKET WITH 2.7M LEAD & PLUG



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Epson: HP LaserJet, Postscript printers and others. Z5442 (6.25" version) Z5443 (3.5"version) £14.96 each

# AMSTRAD PRINTER **MECHANISM**



X9013 Brand spanking new mech, but not sure which model printer this was intended for. The Amstrad part no is Z70315 It's virtually complete, just the print head missing. ideal for modellers to strip down - there's a wealth of mechanical bits and pieces + 2 24V, 24 position stepper motors and an opto



# **Electronics Template**



Order Code H80



## **MOTORS - BATTERY 1V-12V**

3 Different Model Motors, £1, Order Ref: 35. pin to Start 3V DC Motors for model aircraft etc, 5 for £1. Order Ref: 134

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Ref: 643. 12V Motor ex BSR record player, £1, Order Ref: 687 9V Cassette Motor, brushless, £1.50, Order Ref:

1.5P14 4aHP 12V D.C. Motor, Smiths, £6, Order Ref; 6P47. 4HP 12V D.C. Motor, Smiths, £8, Order Ref: 8P54

4HP 12V D.C. Motor, Smiths, £10, Order Ref: 10P125.

## MAINS MOTORS WITH GEARBOXES

5 r.p.m. 60W, £5, Order Ref: 5P54 110 r.p.m. 60W, £5, Order Ref: 5P172. 150 r.p.m. 60W, £8, Order Ref: 8P57. 200 r.p.m. 60W, £8, Order Ref: 8P58

1 Rev Per 24hrs 2W Motor, £1, Order Ref: 89. 1 Rev Per 4hrs 2W Motor, £1, Order Ref: 2P239 1 Ray Per Hour 2W Extra Small Motor, 2 for £1, Order

Ref: 500. 12 r.p.h. Motor, £2, Order Ref: 2P342. 20 r.p.h. Motor, £1, Order Ref: 1010. 4 r.p.m. 2W Motor, £2, Order Ref: 2P346. 4 r.p.m. 2W Motor, £2, Order Ref: 2P393. 15 r.p.m. 2W Motor, £2, Order Ref: 2P321. 200 r.p.m. 2W Motor, £1, Order Ref: 175. 250 r.p.m. 2W Motor, £1, Order Ref: 750.

## **MAINS MOTORS**

3/4 Stack Motor with 1/4" spindle, £1, Order Ref: 85. Stack Motor 11/2" with good length spindle from each side, £2, Order Ref: 2P55.
Stack Motor 1'/4" with 4" long spindle, £2, Order Ref:

Motor by Crompton 0:06HP but little soiled, £3, Order Ref: 3P4

JAP Made Precision Motor balanced rotor reversible. 1500r.p.m., £2, Order Ref: 2P12.

EMI 2-speed and reversible, £4, Order Ref: 4P80. Very Powerful Mains Motor with extra long (21/2") shafts extending out each side. Makes it ideal for a reversing arrangement for, as you know, shaded-pole motors are not reversible, £3, Order Ref: 3P157.

## **MOTORS - STEPPER**

Mini Motor by Philips 12V-7-5 degree step, quite standard, data supplied, only £1, Order Ref: 910. Medium Powered Jap made 1.5 degree step, £3, Order Ref: 3P162.

Very Powerful Motor by American Philips, 10V-14V 7:5 degree step, £10, Order Ref: 10P128.

## **MAINS TRANSFORMERS**

5V 45A, £20, Order Ref: 20P16. 6V 1A, 2 for £1, Order Ref: 9. 8V 1A, £1, Order Ref; 212 9V 1/4A, 2 for £1, Order Ref: 266. 9V 1A. £1. Order Ref: 236. 10V 1A, £1, Order Ref: 492 12V %A. 2 for £1, Order Ref: 10. 12V 1A, £1, Order Ref: 436 12V 2A £2, Order Ref: 2P337. 15V 1A, £1, Order Ref: 267. 17V 1A. £1. Order Ref: 492. 18V 1/2 A, £1, Order Ref: 491 20V 4A, £3, Order Ref; 3P106. 24V 1/2 A, £1, Order Ref: 337 28V 21/2 A. £4, Order Ref: 4P24. 36V 3A, £3, Order Ref: 3P14 40V 2A, £3, Order Ref; 3P107 43V 31/2A, £4, Order Ref: 4P14. 50V 2A fully shrouded, £5, Order Ref: 5P210. 50V 15A, £20, Order Ref: 20P2. 90V 1A, £4, Order Ref; 4P39. 675V 100mA, £5, Order Ref: 5P166. 3kV 3mA, £7, Order Ref: 7P7. 4kV 2mA, £5, Order Ref: 5P139 6V-0-8V 10VA, £1, Order Ref: 281. 9V-0-9V 5VA, £1, Order Ref: 661. 12V-0-12V 2V 3VA, £1, Order Ref: 636. 12V-0-12V 6VA, £1, Order Ref: 811. 12V-0-12V 40VA, £3.50, Order Ref: 3.5P7. 15V-0-15V 1VA, £1, Order Ref: 937. 15V-0-15V 15VA, £2, Order Ref: 2P68. 18V-0-18V 10VA, £1, Order Ref: 813. 20V-0-20V 10VA, £1, Order Ref: 812. 20V-0-20V 10VA, £2, Order Ref: 2P85 20V-0-20V 20VA, £2, Order Ref: 2P138. 20V-0-20V 80VA, £4, Order Ref: 4P36. 38V-0-38V 20VA, £2, Order Ref: 2P156. 90V-0-90V 100VA, £4, Order Ref: 4P39.

## SPECIAL TRANSFORMERS

15VA gives 1V, 7V, 8V, 9V or 10V, £1, Order Ref: 744. 8V + 8V 10VA, £15, Order Ref: 15P51 38V-0-36V 15VA with regulator winding, £10, Order Ref: 10P36

230V-115V auto transformer 100VA, £2, Order Ref:

Ditto but 10VA, £1, Order Ref: 822. Ditto but 250VA, £3, Order Ref: 3P142

# MISCELLANEOUS BARGAINS £1 BARGAIN PACKS

# ALMOST ALL OF THE BARGAINS OFFERED LAST MONTH ARE STILL AVAILABLE. IF IN DOUBT. GIVE US A RING.

0V-20V DC Panel Meter. This is a nice size 65mm sq. It is ideal if you are making a voltage variable instrument or battery charger. Price £3, Order Ref; 3P188.

Flashing Beecon. Ideal for putting on a van, a tractor or any vehicle that should always be seen. Uses a XENON tube and has an amber coloured dome. Separate fixing base is included so unit can be put away if desirable. Price £7.50, Order Ref: 7.5P13.

## 12V 2A Transformer, £2, Order Ref: 2P337.

Another 12V-0V-12V Transformer is a 50VA and is suitable for dropping through the chassis or as it is fitted with four pillars it can be mounted above the chassis. Also should you want a 12V 4A transformer then this one should be quite suitable, you use just one half of the secondary. Price £3.50, Order Ref: 3.5P7.

High Resolution Monitor. 9" by Philips, in metal frame for easy mounting. Brand new, offered at less than the price of the tube alone, £15. Order Ref:15P1.

15W 8" 8 Ohm Speaker and 3" Tweeter. Amstrad, made for their high quality music centre, £4 per pair, Order Ref: 4P57.

Insulation Tester with Multimeter. Internally generates voltages which enables you to read insulation directly in megohms. The multiinsulation directly in megohms. meter has four ranges, AC/DC volts, 3 ranges milliamps, 3 ranges resistance and 5 amp range. These instruments are ex-British Telecom but in very good condition, tested and guaranteed OK, probably cost at least £50, yours for only £7.50 with leads, carrying case £2 extra, Order Ref: 7.5P4.

We Have Some of the above testers but slightly faulty, not working on all ranges, should be repairable, we supply diagram, £3, Order Ref:

250W Light Dimmer, Will fit in place of normal wall switch, only £2 each, Order Ref: 2P380. Note these are red, blue, green or yellow but will take emulsion to suit the colour of your room. Please state colour required.

LCD 31/2 Digit Panel Meter. This is a multirange voltmeter/ammeter using the A-D converter chip 7106 to provide five ranges each of volts and amps. Supplied with full data sheet. Special snip price of £12. Order Ref: 12P19.

Multi Tester. 19 ranges, ex-British Telecom reconditioned. These measure AC and DC volts, DC milliamps and have three resistance ranges made to BT specification and 20,000 opv movement. Complete with test prods, £8.50, Order Ref: 8.5P3. Carrying case with handle £2 extra.

Speed Controller, Suitable for the C5 or other DV 12V motor. Complete kit £18, 18P8, already made, £29.50, Order Ref: 29.5P2.

Clock Module. 2" LCD display, requires 1.5V battery, goes back to zero when switched off so ideal for timing operations, £2, Order Ref: 2P307

Mini Blow Heater. 1kW, ideal for under desk or airing cupboard, etc. Needs only a simple mounting frame, £5, Order Ref: 5P23.

Medicine Cupboard Alarm. Or it could be used to warn when any cupboard door is opened. The light shining on the unit makes the bell ring. Complete built and neatly cased, requires only a battery, £3, Order Ref. 3P155.

Don't Let it Overflow. Be it bath, sink cellar. sump or any other thing that could flood. This device will tell you when the water has risen to the pre-set level. Adjustable over quite a useful range. Neatly case for wall mounting, ready to work when battery fitted, £3, Order Ref:

Quartz Clock Mechanisms. Complete with 2 sets of hands, modern or period, made up ready to work, £3, Order Ref: 3P111.

Amstrad Modern FM240. In as new condition but customer return, so you may need to fault find, £8, Order Ref: 6P34.

This is the £1 Bargain Packs List 2 - watch out for lists 3 and 4 next month

3 x Battery Model Motors, tiny, medium and large. Order Ref: 35.

2 x Tuning Condensers for medium wave radios. Order Ref: 36.

Miniature 12V Relay with low current consuming coil, 2 x 3A changeover contacts. Order Ref: 51. 2 x Ferrite Slab Aerials with medium wave coils. Ideal

for building small radio. Order Ref: 61.

2 x 25W 6 OHM Variable Resistors, Ideal for loud speaker volume control. Order Ref: 69.

2 x Wire Wound Variable Resistors in any of the following values, 18, 35, 50, 100 ohms, your choice Order Ref: 71.

4 x 30A Porcetain Fuse Holders. Make your own fuse board. Order Ref: 82. 2 x 65" Meal Fan Blades for 5% shaft. Order Ref:

Mains Motor to suit the 6 "2" blades. Order Ref: 88. 1 x 4.5V 150mA DC Power Supply. Fully enclosed so quite safe. Order Ref: 104.

10 each red and black small size Crocodile Clins. Order Ref: 116.

15mm Twin Wire, screened. Order Ref: 122A 100 Plastic Headed Cable Clips, nail in type, several

sizes, Order Ref: 123. 4 x MES Batten Holders Order Ref: 126.

Complete Pocket Size MW Radio, believed OK but not d. Order Ref:133R.

4 x 2 Circuit Micro Switches (Licon), Order Ref: 157. 1 x 13A Switch Socket, quite standard but coloured. Order Ref: 164.

1 x 30A Panel Mounting Toggle Switch, double pole, Order Ref: 166. 2 x Neon Numicator Tubes. Order Ref: 170.

100 x % Rubber Grommets. Order Ref: 181. 8 x BC Lamp Holder Adaptors. Order Ref: 191. 8 x Superior Type Push Switches. Make your own key-board. Order Ref: 201.

Mains Transformer 8V-0-8V 1/2A. Order Ref: 212. 2 x Sub Min Toggle Switches. Order Ref: 214. High Power 3" Speaker (11W 8ohm). Order Ref: 246. Medium Wave Premeability Tuner. Its almost a complete radio with circuit. Order Ref; 247.

6 x Screwdown Terminals with through panel in-sulators. Order Ref: 264.

LCD Clock Display. '2" figures. Order Ref: 329.

10 x Push On Long Shafted Knobs for '4" spindle. Order Ref: 339.

2 x ex-GPO Speaker Inserts, ref 4T. Order Ref: 352. 100 x Sub Min 1F Transformers. Just right if you want coil formers. Order Ref: 360. 1 x 24V 200mA PSU. Order Ref: 393.

1 x Heating Element, mains voltage 100W, brass ened. Order Ref: 8.

1 x Mains interference Suppressor. Order Ref: 21 3 x Rocker Switches, 13A mains voltage. Order Ref:

1 x Mini Uni Selector with diagram for electronic jigsaw. Order Ref: 56.

x Appliance Thermostats, adjustable up to 15A. Order Ref: 65. x Mains Motor with gearbox giving 1 rev per 24 hrs.

Order Ref: 89. 10 x Round Pointer Knobs, for flatted %" spindles.

Order Ref: 295.

1 x Caramic Wave Change Switch, 12 pole, 3 way with \u03c4" spindle. Order Ref: 303. 1 x Tublar Hand Mike, suits cassette recorders, etc. Order Ref: 305.

2 x Plastic Stethosets, take crystal or magentic in-serts. Order Ref: 331.

20 x Pre-set Resistors, various types and values. Order Ref: 332.

6 x Car Type Rocker Switches, assorted. Order Ref: 333.
10 x Long Shafted Knobs fpr ¼" flatted spindles. Order Ref: 339.

1 x Reversing Switch, 20A double pole or 40A single pole. Order Ref: 343. 4 x Skirted Control Knobs, engraved 0-10. Order Ref:

3 x Luminous Rocker Switches. Order Ref: 373

2 x 1000W Tublar Heating Elements with terminal ends, Order Ref: 376.

1 x Mains Transformer Operated Niced Charger, cased with leads. Order Ref: 385. 2 x Clockwork Motors, run for one hour. Order Ref: 389.

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K2100 The total cost of a complete set of all components to build this unit is £126.37. Our special discount price for all parts bought together as a kit K2100SA Series Audiophile, with extra selected components. £112.46

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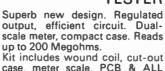
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INCORPORATING ELECTRONICS MONTHLY

**VOL. 24 No. 8** 

**AUGUST'95** 

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# Constructional Project

# SOLAR SEEKER

# JIM and CHRIS WILES

A solar battery charger with a moving panel that tracks the sun.

ARE for the environment has become more and more important to everyone and quite a bit of commercial attention is now placed on the manner in which electricity can be produced from renewable sources. An example of this can be seen from the arrays of wind powered generators which are now obvious in the West Country.

To take an interest in renewable energy sources, though, it is not necessary to manufacture a large whizzing machine with forty-foot blades and place this in the garden. As in the early days of amateur radio, in some ways it is still possible for electronics enthusiasts to move technology forward. In a modest way, this Solar Seeker Battery Charger strives to do that. Its purpose is to get more energy from the available sunlight by tilting its solar panel to follow the sun.

# SOLAR BATTERY CHARGERS

The simplest form of solar-powered battery charger circuit is shown in Fig. 1. In this circuit, the voltage generated at the terminals of the solar panel is used to charge the battery via current limiting resistor R1. The resistor prevents batteries from being recharged at too high a rate for too long. It can be omitted if the solar panel used has a low current rating. Diode D1 prevents the battery from discharging through the solar panel when sunlight is absent.

A graph of charging current plotted against time of day is shown in Fig. 2. It was plotted using measurements gathered in the West of England in July, using a

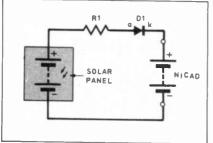


Fig. 1. Simple solar powered battery charger circuit.

10cm by 12cm solar panel. The curves assume that an 8.4V NiCad battery is being charged. Curve 1 shows the charging current from a panel laid flat on the ground. Curve 2 shows the charging current available if the panel is tilted and moved by hand to follow the sun.

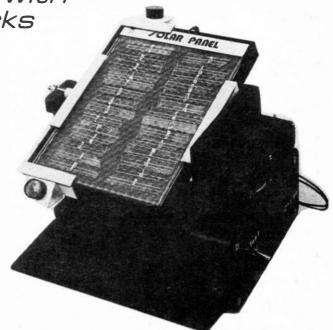
From the graphs it can be seen that not only does the tilted panel have a higher prospective charging current for most of the day, but that it can provide the current for several hours longer than a static panel. Calculating the

total charge which can be achieved in a day, it becomes apparent that a tilting panel can theoretically achieve as much as 100 per cent more energy capture than a static panel.

With a tilted solar panel the designers have found that it is also possible to achieve significant charging levels to within 20 minutes of sunset. They assume that the

same is true of sunrise, although they claim to remain firmly asleep for most of these!

Obviously, a solar panel which follows the sun generally has a greater available energy capture and therefore it seemed a good idea to try to automate its tracking movement. The problem that had to be overcome, though, was how to do this without consuming a lot of energy.



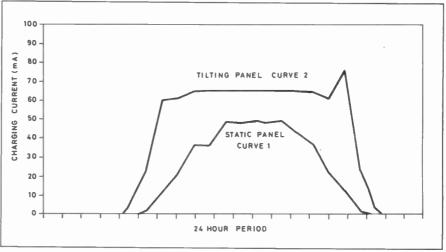


Fig. 2. Measured solar panel performance.

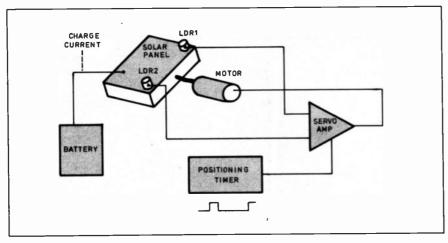


Fig. 3. Block diagram for the Solar Seeker.

# SEEKING APPROACH

The system adopted is shown in Fig. 3. Its concept is simple. The solar panel is rotated about a horizontal axis by a small electric motor. Two light dependent resistors (l.d.r.s) are mounted on the solar panel. These sense the position of the sun with respect to the position of the panel and any difference is converted into a small drive voltage, which is amplified and drives the motor to correctly reposition the panel.

Output current from the solar panel charges a battery and this battery is also used to power the positioning circuit. As even small motors can draw quite large currents, the positioning circuit must be switched off for most of the time. If this is not done then the battery will probably never charge or even be totally flattened by the permanent load of the motor.

In the Solar Seeker, a low power timer is used to control the power supply to the motor and positioning need only occur for about five seconds in every 30 minutes.

The system shown is totally self-powered. In very bright sunlight it was found that the solar panel can power the positioning circuit almost on its own. In less bright conditions, the battery is used briefly to power the positioning process and then return to accepting charge.

In principle, the system can be used to charge a number of NiCad cells arranged in series as a battery. The only requirement is that the battery should provide a high enough voltage to adequately drive the motor via the positioning circuit. For the Solar Seeker as described, the battery is assumed to be a NiCad PP3, although guidance will be given on component changes to enable the circuit to charge AA or AAA cells arranged as a six-cell battery.

# **POSITIONING**

The l.d.r.s are attached to the solar panel at an angle of approximately 45 degrees. This angle ensures that each sensor receives the same amount of sunlight only when the panel is pointed directly at the sun. When the panel is not directly pointed at the sun, each l.d.r. receives a different amount of light and this difference determines how much repositioning the panel requires.

The circuit used to achieve automatic positioning is shown in Fig. 4. The two l.d.r.s are labelled as R6 and R7. Facing straight into bright sunlight, the l.d.r.s have a resistance of about 50 ohms. When angled slightly away from the sun the

resistance rises slightly to 55 or 60 ohms. In the latter condition, a difference of about 30mV appears at the input to op.amp IC2a. The voltage difference is amplified by IC2a to produce a 2V drive to the motor M1. Preset potentiometer VRI is used to offset any mismatch in the l.d.r.s and their mounting angles.

Power to drive the positioning motor is provided by IC2a, which is an L272 low-cost dual power op.amp. This is capable of operating at supply voltages down to 4V, has a 1A drive capability and a power rating of 1W. The suggested motor draws a maximum of about 50mA and at this current the low voltage supply rails ensure that the chip dissipates less than the maximum power rating. A heatsink is not required.

voltage of the main supply rail. Resistor R12 is included to limit the maximum current drawn by the motor and reduce battery drain.

IC2 needs to be protected from inductive spikes created by the motor. Diodes D7 to D10 ensure that such transients cannot appreciably exceed the supply rails. However, even when clamped by the diodes, these transients can, in some situations, produce radio frequency interference (RFI). This was tested for with the motor type used on the prototype and no RFI was detectable by a radio located a metre from the Solar Seeker.

It should be borne in mind, though, that RFI may become appreciable and require alleviating measures if the design is varied and larger panels needing bigger motors are used.

The flashing light emitting diode (l.e.d.), D11, operates when the positioning circuit is energised and the panel is being moved. Although this l.e.d. does not actually require a series resistor, R15 has been included to minimise current drain.

# TIMER CIRCUIT

The timer circuit, which triggers the tilting system, has to remain dormant for 30 minutes, during which time it must consume as little current as possible. To achieve this, a CMOS 555 timer chip (IC1) has been used: it consumes only 100 microamps when in the "off" state.

The circuit is shown in Fig. 5 and is configured as a conventional astable. It is important that the timing capacitor, C1, is of the tantalum variety. Such capacitors have a very low current

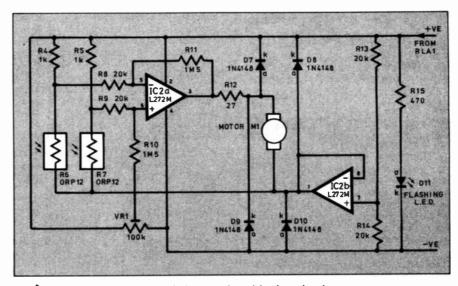


Fig. 4. Solar panel positioning circuit.

The gain of IC2a is set by the ratio of resistors R11/R8 and R10/R9 and is nominally 75. This gain value is suitable for use with the modified radio control (RC) servo motor used in the prototype, but it may need to be varied if a different motor is used.

In order to track the sun, the panel must move both forwards and backwards. Consequently, the drive voltage to the motor needs to be bidirectional, and so a low impedance centre power rail is required. This is achieved by using the second op.amp of the L272, IC2b.

Operating as a unity gain buffer, IC2b is biased via its non-inverting input, pin 7, by resistors R13 and R14 to produce an output on its pin 1 which is at half the

leakage, which is essential in this circuit; the use of an electrolytic capacitor is not recommended.

The 20 megohm resistor, R1, results in a charging current to C1 of between 0.5 to 1.0 microamps. Although this is less than the nominal leakage currents for 100µ tantalum capacitors, and in theory the capacitor should not charge up fully, the designers tried a number of 100µ tantalum capacitors and found that the circuit worked with all of them.

The timer output, IC1 pin 3, has a high current-sink capability but acts as a poor source. Therefore the output from IC1 is arranged to switch on the supply to the positioning circuit when its output is in the "off" state (0V).

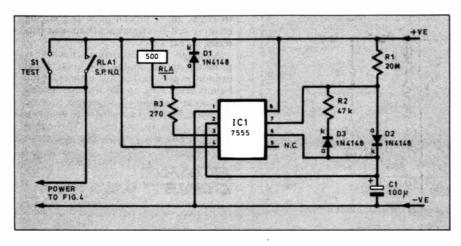


Fig. 5. Circuit diagram for the timer. It is important that C1 is a tantalum type. The relay is a reed type.

The circuit includes two diodes, D2 and D3, which are normally associated with low duty cycle operation (output off for a long time and on for a short time). However, it was found that although the duty cycle was the reverse of this, the circuit was erratic without the diodes. With them included, though, repeatability was within a few seconds in thirty minutes.

Switching of the supply voltage to the positioning circuit is achieved by reed relay RLAI. Switch SI has been included to bypass the relay, to enable testing and initial positioning of the solar panel.

# MOTORING

The choice of motor and transmission method caused the greatest amount of thought in this project. The difficulty lies in the transmission rather than the motor. This is because the object was to produce a speed of six r.p.m. from a 2V or 3V drive voltage whilst achieving low current drain.

To achieve this, an efficient motor with a compact gearbox is required. Small motors are readily available but compatible compact gear boxes are not the easiest things to find. Industrial grade servo motors would fit the bill, but are not cheap! After a lot of

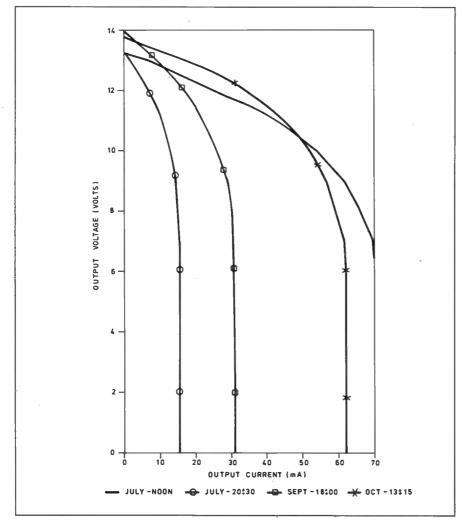


Fig. 6. Voltage and current characteristics of the solar panel measured on four occassions.

head scratching, the simple approach was adopted: adapting a readily available radio control (RC) servo motor.

These motors already include a high reduction gearbox, but they are designed to be driven by a pulse code signal, rather than d.c. power. The solution is to open up the servo and connect wires directly to the internally mounted motor, discarding the small pulse decoding circuit board.

Via its gearbox, the modified motor is coupled to turn the solar panel, consuming an average of 10mA, with an instantaneous peak demand of about 50mA.

# SOLAR PANEL

Solar panels are constructed from a number of silicon solar cells wired in series. Each cell will produce a little more than 0.4V in bright sunlight and the current available is proportional to the area of each cell. The ready-made panel used in this project has 28 cells connected to produce a nominal 12V at 30mA.

The measured characteristic of the solar panel is shown in Fig. 6. The highest voltage is produced when unloaded. As loading is increased, the voltage drops and eventually the panel's output current saturates at a level dependent upon the sunlight intensity. It can be seen from the graphs that the panel has a good output level, not just in summer, but also as winter approaches.

The saturated-current portion of the curves is suited to automatic NiCad constant current charging without additional circuitry. There are two problems, though. Firstly, sunlight intensity varies throughout the day and so a predictable constant current cannot be guaranteed, despite automatically tracking the sun's direction.

The second problem is that in bright sunlight, the charging rate may be higher than required for a nominal 16-hour charge. Care is then needed to ensure that current is discontinued when the battery is fully charged. The problem can be solved with a suitable charging circuit, but first it is necessary to know the acceptable charging range for the battery concerned.

# **BATTERIES**

The panel positioning and timing circuitry can position and charge PP3 batteries, or, either six AAA or AA cells in an appropriate holder. The limit of operation is reached with only four cells, which barely tilt the panel.

The maximum charge currents for the six-cell combinations and the PP3 are shown in Table 1. All but the PP3 can be fast charged. As the suggested panel maximum current (about 70mA) exceeds the maximum charge rate for the PP3 and AAA, cells any charging circuit must act to limit the charge current appropriately. If fast charging is employed, it is also necessary to reduce the charge rate when the battery is fully charged.

For the larger AA cells and the suggested solar panel, it is almost possible to dispense with any control over the charging, simply remembering to disconnect after, say, 16 hours: this is because the maximum current from the panel is close to the standard charge rate. It should be noted that, on the brightest summer days, only the AA cell combination is capable of accepting all the incident solar energy.

In general terms, it is not a good idea to charge NiCad cells as a battery of several cells. For the PP3, this cannot be avoided as the connections between cells are internal. If several individual cells are to be

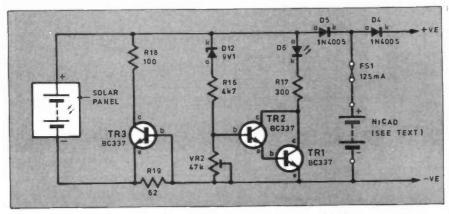


Fig. 7. Shunt regulator battery charging circuit.

charged in series as a battery, then it is good practice to individually discharge the cells, but to no less than IV per cell.

# CHARGING CIRCUIT

The charging circuit is shown in Fig. 7. As a basic design it can cope with charging either a PP3 or the six-cell combinations listed in Table 1. The component values shown in Fig. 7 are for charging a PP3.

Two purposes are served by the circuit: prevention of excessive charging currents and illumination of an l.e.d. as an approximate indication of charge completion. The current illuminating the l.e.d. reduces the available current to the battery and can be arranged to terminate fast charging.

Basically, the circuit consists of two shunt regulators. It has been designed on the principle that all energy should pass unimpeded to the battery unless it is unsafe to do so. Safe charging current is regulated by transistor TR3, and is set by the value of resistor R19.

As the current through R19 approaches the safe charging current ( $I_c$ ) the voltage between the base (b) and emitter (e) of TR3 rises to 0.6V and the transistor starts to turn on. This shunts current away from the charging circuit and dissipates the surplus energy in R18. The value of R19 is calculated from: R19 = 0.6 /  $I_c$ .

The circuit around transistors TR1 and TR2 serves to warn the user and reduce the charge current as the NiCad voltage rises. The battery voltage is sensed by Zener diode D12, resistor R16 and preset VR2. The circuit values were chosen to minimise current drain.

Preset VR2 is adjusted so that resistor R17 and l.e.d. D6 shunt the required current when the battery reaches its full charge. This can be arranged to either terminate charging or to reduce charging so that the solar panel can continue safely charging the NiCad at its standard rate.

In this circuit, the current is dumped through the l.e.d., which glows at its brightest when the battery is fully charged. Note that the current to this part of the circuit is ultimately limited by TR3 and if all the available current is dumped through the l.e.d. then no current will pass to the battery.

Typical values for resistor R19 and Zener diode D12 to suit different batteries are given in Table 2. Two different modes of charging are illustrated by the values in this table. The PP3 is charged at its standard rate and the current flow virtually ceases as it becomes fully charged.

On the other hand, the six AAA cells are fast-charged at their maximum rate and then charging is reduced to the standard rate. The details in Fig. 7 and Tables 1 and 2 should enable experienced constructors

to extend the approach to cover charging of other cell arrangements (lead acid batteries, for example).

Although the positioning circuitry is protected by diode D4, the solar panel cannot be protected in this manner. If the battery is reversed, an effective short circuit will occur and the consequences may be dramatic. Fuse FS1 is included to minimise this risk. Its value may need to be changed if the motor differs from the RC servo recommended.

# BOARD CONSTRUCTION

The complete circuit is constructed on a single piece of stripboard, whose assembly and track-cutting details are shown in Fig. 8. Assemble the components in any order most convenient. Sockets should be used for both i.c.s. Note that resistor R1 is made up of two resistors connected in series. Solder pins can be used to make wiring to off-board components easier. With the compact size of the unit the connections should be sleeved.

All stripboard tracks should be checked for solder bridges before the circuit is powered-up. In particular, the solder joints and tracks around the timer IC1 should be checked and cleaned to minimise track-to-track leakage currents. Make sure that capacitor C1 is correctly orientated. It is also essential to observe the normal handling precautions for CMOS devices, ensuring that you discharge static electricity from yourself before handling them.

The stripboard is housed in a small plastic box. The switch and l.e.d.s are mounted on the lid. For the triangular l.e.d. D6, a small hole was first drilled in the box and a triangular needle file then used to carefully cut the correct shape.

As the plastic box is fastened to the Solar Seeker stand with screws, make sure to use a strip of tape to insulate the screw heads, before finally positioning the circuit board.

# MOTOR MODIFYING

To modify the RC servo motor, the motor is carefully opened up (usually by

**TABLE 1: Battery charging information** 

Battery Type	Cell Qty.	Nominal Voltage (1)		Charge Standard mA		Cha Effici (2)	-
PP3	7	8-4V	110	11	11	72%	
AAA x 6	6	7-2V	180	18	45	60%	70%
AA x 6	6	7·2V	500	50	500	60%	90%

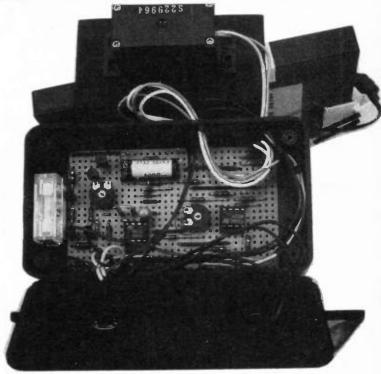
- (1) Typical operating voltages lie in the range 1-24V to 1-27V per cell.
- (2) Energy stored / Energy supplied. At the standard charge rate

(3) Indicative efficiency at the fast charge rate.

On charge, the maximum voltage per cell must not rise above 1.5V. The minimum voltage per cell on discharge must not fall below 1V.

TABLE 2. Essential factors for charging various batteries

			Jacteri	-		
Battery	<b>R</b> 19	D12	D6	Charge Current	Residual Charge	Charge Time
			Max.		Current	
	ohms	value	mA	mA	mA	hrs
PP3	62	9V1	10	10	<1	14
AAA x 6	15	6V8	22	40	18	5
AA x 6	12	6V8	22	50	28	16



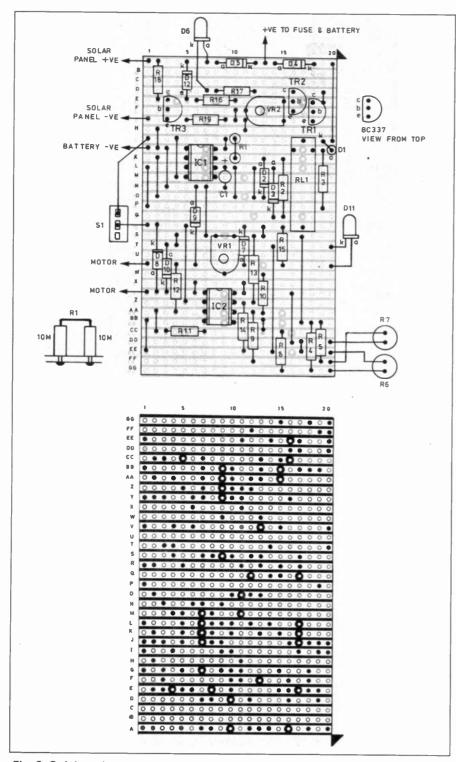
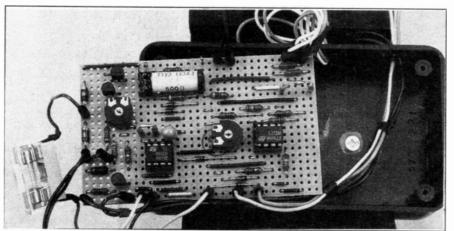


Fig. 8. Stripboard component layout, wiring and track cutting details for the Solar Seeker. The completed circuit board is shown below.



# COMPONENTS

Resistor	5
R1 R2	10M (2 off – see text) 47k
R3 R4, R5	
R6, R7	ORP12 light dependent resistor (2 off*)
R10, R11 R12	20k.1% (2 off) 1M5.1% (2 off) <b>See</b> 27Ω 0.6W
R13, R14	27Ω 0·6W SHOP 20k 1% (2 off) <b>TALK</b>
R15 R16	470Ω 4k7
R17	300Ω 0·6W (see text and Table 2)
R18 R19	$100\Omega \ 0.6W$ $62\Omega \ 1\%$ (see text and
All 0-25W	Table 2) 5% carbon film or better,
inless othe	rwise specified. should both be purchased at
Potentio VR1 VR2	meters 100k submin. preset horiz. 47k submin. preset horiz.
Capacito	r
C1	100μ tantalum, 10V

# Semiconductors

tl

D1 to D3, D7 to D10 1N4148 signal diode (7 off 1 N4005 rectifier diode D4, D5 (2 off) I.e.d., triangular I.e.d., flashing D6 D11 D12 9V1 Zener diode 400mW (see text) TR1 to TR3 BC337 npn transistor (3 off) IC1 ICM7555IPA CMOS IC2 L272M dual power

# Miscellaneous

S1 s.p.s.t. min. toggle switch 125mA fuse and 20mm fuseholder RLA1 Submin. s.p. reed relay, 500 ohm coil

op.amp

Solar panel, 98mm x 128mm, 12V/30mA output at 100mW/sq.cm illumination; RC servo motor with mounting hardware (see text); stripboard 0.1 inch grid, 33 strips x 20 holes; plastic case 115mm x 60mm x 30mm; test resistor, 47 ohms 5% 0·25W carbon film; plywood and beading for frame (see Fig. 9); nuts, bolts, screws etc; A4 slide binder; adhesive; 1mm solder pins; connecting wire; solder, etc.

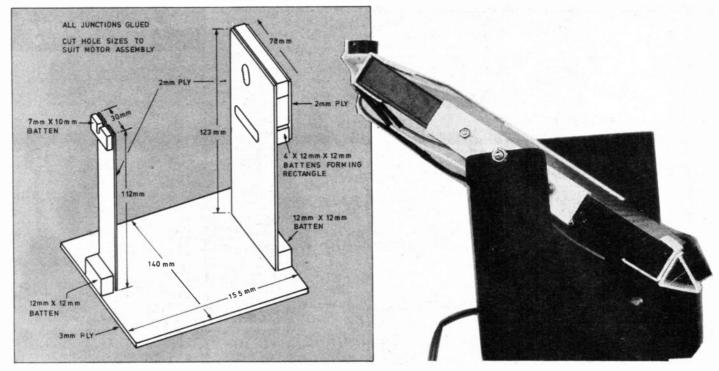
Approx cost guidance only

£30

undoing four screws) and the internal circuit board exposed. It was found with the prototype that the gearbox was in a separate compartment from the circuit board, but that both compartments shared the same screws for fastening.

If the joins in the case are looked for, it should be possible to access the circuit board and motor connections without spilling all the cogs on the floor! The join furthest from the front (drive shaft end) of the servo is the one to carefully prise first.

The motor's circuit board was disconnected from the internal potentiometer and motor and put aside for another project. It may be easiest to snip the connecting wires



servo motor body is screwed. The solar

panel fits between the two uprights and is supported by two "RC servo horns" (small plastic couplings available from RC model

shops) which pass directly through holes in

the uprights. One of the horns is push-fitted

directly to the motor. The other has an M3

bolt passed through it and locked with a

(c)

nut, to form an axle.

Fig. 9. Mounting frame for the solar panel assembly.

and use them as connection points. Two wires are soldered to the motor leads or terminals and brought out from the case.

Sometimes, an RC servo may have interference suppressing components connected to the motor terminals and case. These may be small capacitors or resistors; it should do no harm to leave them in place.

# **MECHANICS**

Almost the entire hardware construction of the Solar Seeker frame can be made from modelling plywood and square beading. Details of the frame are shown in Fig. 9, Fig. 10 and the photographs.

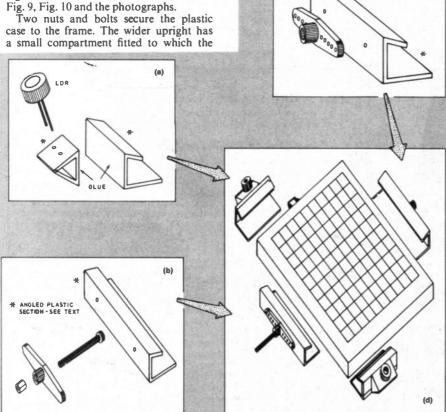


Fig. 10(a). L.D.R. assembly, (b) axel mounting, (c) servo coupling, (d) solar panel assembly.

Both the l.d.r.s and the servo horns are mounted on the solar panel using a plastic "slide binder". This is a long clip, with an almost triangular section, which is normally used to hold together several sheets of A4 paper. It is cut into various lengths and used as shown in Fig. 10.

The panel is assembled with its various clips and trailing wires, and then mounted onto its stand. The servo motor is pushfitted on one side and the bolt is slid into a "U" shaped slot on the other (appropriate tubing can be obtained from model shops and cut up to form bearings if needed).

Cables from the solar panel and l.d.r.s pass through the slot in the wider upright and connect to the circuit case. A slot is used rather than a hole to allow for movement in the cables as the panel turns. After testing, the cables can be secured to the panel using small adhesive clips.

# SETTING-UP

To set up the charger circuit, in place of the solar panel connect a 12V d.c. source (e.g. battery or power supply) to the board via a 47 ohm resistor. Rotate preset VR2 until l.e.d. D6 turns on. Now rotate VR2 to the opposite end of its track and the l.e.d. should turn off. Take a fully charged PP3 NiCad (or the combined AA or AAA cells as appropriate) and connect to the charger circuit via fuse FS1. The fuseholder is not mounted in the case but simply connected externally between the unit and the battery.

Using a multimeter, check the charge current going into the battery: it should be the required safe charging current (see Table 2) with the l.e.d. off. Leave the battery connected and charging for five minutes, and then adjust VR2 until the l.e.d. just glows at its brightest. For this setting to be most accurate the battery should have been charged at the same current used in the charge circuit. Ultimately, this setting can be checked using a battery which has already been charged using the Solar Seeker.

With the timer circuit, check that l.e.d. D11 flashes when S1 is closed. Open the switch and wait. D11 should now flash for five seconds every 30 minutes. The first

timed period may be longer than 30 minutes, as the circuit initialises itself, but after that a very repeatable sequence should be apparent.

In the unlikely event that trouble is experienced with the timer, test the circuit with a 100k resistor connected across R1. This should allow D11 to flash every ten seconds. The most likely cause of failure with a high value for resistor R1 is leakage current between adjacent tracks. Re-clean between the tracks around IC1. If difficulties persist it may be necessary to reduce the value of R1. Alternatively, simply try replacing C1 in case its leakage current is greater than expected.

# LIGHTLY TEST

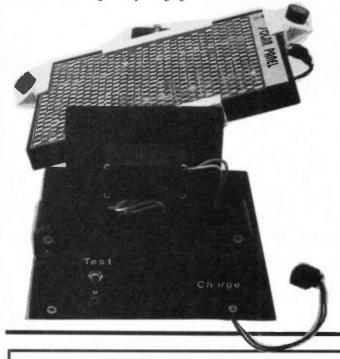
When the solar panel, motor and electronics have been mounted on the frame, rotate the solar panel by hand to ensure that the motor allows movement over the full range. If the motor was at the extreme of its travel when the panel was fitted then remove it, rotate the panel, and refit to ensure proper movement. On test, it may be necessary to swap over the motor connections to obtain the correct drive sense. Connect up the battery and switch on \$1.

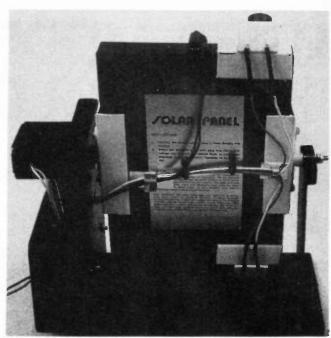
A very bright light source, such as a desk light about 200mm away, is needed to check that the Solar Seeker tracks correctly. Normal room lights are not a good check. The problem with weak light sources is that they cause much higher resistance values in the l.d.r.s than sunlight, consequently the l.d.r.s do not appear as well matched underly the audition.

matched under such conditions.

Ideally, make any adjustments to VR1 outside in sunlight. However, with the prototype it was found that no adjustment was really needed. It was found, though, that the panel was driven slightly faster in one direction than the other. This was not significant in terms of operation of the solar panel and adjustments of VR1 did not change this.

When finally testing in sunlight, diode D6 should be brightly lit with the battery disconnected and off with it connected. As the battery charges, D6 will gradually come on and grow brighter to give a general indication of charge completion. Switch S1 MUST be turned off during battery charging.





No recommendation is offered about water proofing or permanent exterior installation: the Solar Seeker as described is definitely a fair weather charger. It may be painted, of course, but bring it inside if it rains!

# PERFORMANCE

Using the Solar Seeker, roughly constant charging levels were maintained for longer than had been expected. This can be seen from the shape of curve 2 in Fig. 2, which is surprisingly square. It was concluded that this was due to the fact that the panel voltage drops as the panel heats up.

Any assessment of efficiency is subject to considerations of the weather and time of year. In summer, the design is estimated to use less than one per cent of the energy given to its battery and will gather 20 per cent to 100 per cent more energy than a static panel using a similar charger circuit. The lowest energy gain relates to charging a PP3 and the higher to AA cells or any battery type that can safely accept all the available incident energy.

It is not claimed that the Solar Seeker is an optimum design and the project raises many possibilities and "what if?" questions:

- ★ Is it worth automatically cooling the solar panel?
- ★ Is it worth tilting the panel in two directions?
- ★ Is there a simpler way to tilt the panel?
- ★ What is the optimum charging approach?
- ★ Could camcorder batteries be charged while owners lounge on the beach?
- ★ Could the Solar Seeker benefit Third World countries?

Any of these questions could keep interested experimenters going for some time and many adaptations seem possible.

The Solar Seeker has a wide operating range, but it is unrealistic to expect it to be greatly used in a British mid-winter. Also, unfortunately, it cannot make the clouds go away! It is hoped, though, that you enjoy making the Solar Seeker and that the sun shines often enough for you to make good of use it.

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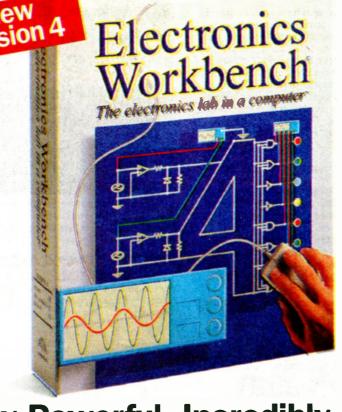
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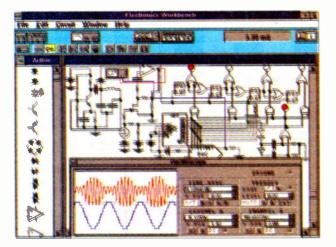
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# Innovations A roundup of the latest Everyday News from the world of electronics

# **COOLING THE WARRIOR**

British cooling system monitor installed in American destroyers

- by Hazel Cavendish

NOTHER important contract has been won for Britain by the small Hampshire firm of Stewart Hughes, whose success in designing a method of detecting mechanical faults in helicopters by analysing vibration signals was reported in the December issue of *EPE*. Their brief to design a small control box to monitor the cooling system of an American Spruance class destroyer was executed with such originality and economy of space that they beat all the competition from United States firms to secure a £650,000 contract. The control box is no larger than an A4 ringbinder ledger.

Maintaining chilled water in new-age Navy ships fully sealed against a nuclear, biological and chemical warfare environment is vital for the functioning of a ship. "The total loss of cooling water would cause the anti-aircraft defence radar to shut down within 20 seconds," says Max Hadley, the Consultant Systems Design Engineer who planned the development of the Intelligent Sensor Interface sold to the US Navv.

What Max Hadley describes as "an experiment on a grand scale" was prompted by the US Navy's drive for economy. The US Defence department asked for tenders from a large number of companies for a demonstration system to be installed in a Spruance class destroyer. It was intended to cut down manpower by using sophisticated equipment to perform monitoring tasks previously done by sailors.

"The result has been a complete Star-Trek type of automated maintenance," says Hadley, "In Hadley, developing our unit, the Intelligent Sensor Interface, we took the opportunity of using some advanced technology. Unlike airborne systems, where one has to rely on proven components, we stuck our necks out and used the latest microchips.

"First we used the new Motorola MC68360 microprocessor, which incorporates both the processor and an Internet interface. The system is designed so that it obeys the TCP-IP networking protocol suite, and if hung on a network connected to the Internet it is possible to select any place in the world and learn the temperature there at that moment.

"While the Motorola processor handles the communication and network side of the work, the TMS320C30 digital signal processor deals with the signal analysis side of the software. A Motorola 56AD16 analogue to digital converter interfaces the sensors. Flash memory is used by both processors. It is only by using advanced chips like these that we are able to get the system down to a little box which screws conveniently to the bulkhead."

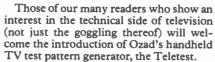


DD978 "Stump", Spruance class destroyer. Photo courtesy of Jane's Information Group

Because Naval ships with their extensive radar installations prove appalling environments for electromagnetic systems, fibre-optic signal cables are used to provide insulation against troublesome interference. "The Falklands campaign taught us that radar blips can play merry hell with electronic systems unless they are proofed against it," says Hadley, who uses the fibre-optic Ethernet as the networking medium.

A second unit called the Diagnostic Processing Unit has also been developed by the company. This acts as an interface between the intelligent sensor and the ship's main network. With the ship's computerised monitoring system a fault can be traced immediately by asking the damage control computer what has gone wrong and where - "just like Star-Trek," says Hadley.





Ozad express the belief that the Teletest is the TV engineer's "Ultimate Toolkit Accessory". Although it has been designed to meet the TV repair engineer's everyday use in the workshop or in the field, the Teletest should prove to be an invaluable test tool for keen electronics enthusiasts as well as AV technicians in companies, colleges and universities.

Teletest provides four essential PAL test patterns: colour bars, grey scale, crosshatch and red purity. A 1kHz audio test tone is also generated. The r.f. video and audio signals are output from a 75 ohmterminated co-axial socket. The instru-ment is pre-tuned to UHF channel 36, with the audio sub-carrier set to 6MHz for

the UK (a 5.5MHz version is available for other countries).

Powered by a 9V battery, Teletest can be connected to the r.f. and line input sockets of TVs and VCRs. An external mains PSU is included for continuous use. Teletest measures 155mm x 80mm x 40mm, weighs 265g and is made from a tough black ABS plastic. It costs £99, excluding VAT and post.

For further information contact Nick Rose at OZAN, 37 Haviland Road, Ferndown Industrial Estate, Poole, Dorset, BH21 7SA. Tel: 01202 877270.



Sevenoaks Adult Education Centre has advised us that they are again offering their successful Radio Amateurs City and Guilds 7650 courses. The next course is for the period 1995/96, full details of which, including cost, dates and times, can be obtained from the Centre, address Bradbourne Road, Sevenoaks, Kent, TN13 3QN, Tel: 01732 451618.



# LE VOXBOX TO THE RESCUE

Jeremy Austin who, together with Guy Dance, designed the Voxbox project (EPE July 1994) has told us of a novel way in which the design is being used. Regular readers will remember that the Voxbox is a tapeless voice recorder capable of recording two messages each of ten seconds duration, or one message of twenty seconds' duration. It was originally produced at GPT's Poole factory for introducing youngsters to a career in hardware engineering.

Voxbox attracted the attention of Poole lifeboat crew member Nigel Burton. He approached Guy Dance with a request for a seagoing version. The RNLI could see the advantage of using a small, self-contained, ultra-simple and reliable voice recorder for capturing distress messages under arduous circumstances at sea.

For their requirement, the RNLI needed to be able to record longer messages, so Guy and the boys back at GPT put in extra circuits to stretch the record time out to two minutes. Seafaring Voxbox is also a tougher cousin of the published version, being water-proof and (almost) sailor proof.

Once completed, this new design lacked only a suitable name before handover could take place. "We decided to call it Le Voxbox," said Guy, "in honour of our French exchange student, David Passaretti, who constructed the final design for us".

Now Le Voxbox has received its sea-trials and Nigel's verdict is that it is a great success. "It isn't practical to keep rewinding a taperecorder," he said. "and when everyone's on deck we can be short-handed for monitoring the radio. Le Voxbox is really very handy, and much better than using pencil and paper!"



Left to right: Nigel Burton, David Passaretti and Guy Dance.

Meanwhile Guy, whose latest exploit with the Voxbox has been to demonstrate its versatility as an auto-dialling burglar alarm, is pondering what will be the next application for his ubiquitous design. "Now Le Voxbox has gone to sea," he said, "we wonder where next it will bob up to the surface?



The Fluke 863 Graphical Multimeter.

# GRAPHICAL MULTIMETER

A graphics facility has been added to Fluke's new 860 series of digital multimeters. This offers users the means of viewing waveforms, graphs, in-situ component signatures, trend plotting and logic activity. The meter also provides, of course, all the functions you would expect on a DMM from a leading manufacturer.

Waveform display mode provides a clear picture of noise, distortion, intermittent failures and glitches, with either a fully automatic or manual set-up. The trend graph plots high resolution meter readings for up to 30 hours, in intervals from one second to 15 minutes.

Meter mode is more traditional, offering both digital and analogue information on a 4.5 digit display and analogue needle graph.

With prices starting at £675 for the 863 model the meters are, however, really only suitable for the serious constructor/designer. For more information contact the Professional Instrument Distributors Association, P.O. Box 35, Skipton, North Yorks, BD23 6PT. Tel: 01756 799737.

# **PLD TRAINER**

New software is one feature included in the latest PLD (Programmable Logic Device) trainer designed and manufactured by E & L Technologies.

Upgrading of this popular training tool has given integral development software based on a version of National Semiconductor's software. Compilation and simulation support is provided for a wide range of devices which include G16V8, G20V8, G22V10, G6001V, MAPL 128, MAPL 144, MAPL 244 and MAPL 268.

Reprogrammable Lattice device 22V10 is supplied with the trainer, and may be programmed from any JEDEC file produced for GAL22V10, regardless of software used to compile the equations.

This PLD trainer has all the features and functions necessary to program, test and use PLDs and is supplied with full supporting documentation. As an educational trainer it is a complete digital electronic system which allows all levels of students to become familiar with and understand the concept and use of PLDs. The price is £450.00, excluding VAT.

For further information contact E & L Technologies, Rackery Lane, Llay, Wrexham, Clwyd, LL12 0PB. Tel: 01978 853920.

# SURFACE MOUNT ADAPTOR

An increasing problem for hobbyist electronics experimenters is that sometimes the ideal chip to use is only available as a surface mounting device (SMD). Whereas experimenters can readily plug various standard dual-in-line (DIL) chips in and out of sockets, the gull-wing configuration of many SMDs precludes their easy use in this way.

CorinTech, though, have introduced a product which should find obvious appeal with experienced hobbyists. Their new range of ceramic SO (small outline) to DIL converters allow a variety of SM i.c.s to be adapted to fit into standard sockets. From SO8 upwards, all standard package styles can now be reconfigured to the equivalent DIL format.

Mike Barton, CorinTech's General Manager, says that "these converters also offer design engineers access to otherwise wasted space". Tracks may be routed and components located both around the SO i.c. and on the underside of the unit. Decoupling capacitors, pull-up/down resistors and protection diodes are easily accommodated and many neat technical solutions can be achieved, particularly in high frequency applications. For further information contact CorinTech Ltd., Ashford Mill, Station Road, Fordingbridge, Hants, SP6 1DG. Tel: 01425 655655.

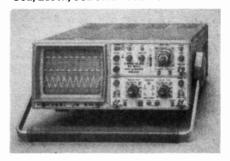
# **NEW SCOPE**

B.K. Electronics are now stocking the new Hameg HM303 oscilloscope. Succeeding the HM203, which sold over 170,000 worldwide, the HM303 bandwidth has been extended to 30MHz, the unit boasts a lighter weight of 5-6kg, and a switched mode power supply. Another

advantage of the scope is that it will work from a mains supply of between 90V and 240V a.c. without adjustment.

The HM303 is priced at £392 plus VAT, and includes two switched probes and a manual. Carriage is free of charge.

For more information contact B.K. Electronics, Unit 1, Comet Way, Southend-on-Sea, Essex, SS2 6TR. Tel: 01702 527572.



# **New Technology** lan Poole looks at a new family of devices which, it is claimed, could be the answer to the r.f. designer's prayers.

HILST the growth of the communications industry is good news for many, it also brings its own problems. The increase in demand for spectrum means that new services have to use higher frequencies. Already many cell phone services are operating at frequencies around 1800MHz. Some services are having to use very much higher frequencies.

# Monolithic M.M.I.C.

To be able to use these frequencies a number of new radio frequency developments have come to the fore. One of these is the monolithic microwave integrated circuit or M.M.I.C.

These come in a variety of forms. There are amplifiers, mixers, dividers and a host of other circuits which have been developed. They are all true r.f. i.c.s, often containing elements like inductors and transmission lines.

Some are capable of operating at exceedingly high frequencies. Fifteen to twenty thousand megahertz (i.e. 15GHz-20GHz) is not unheard of.

There is a wide variety of these circuits which are available for use on a variety of frequencies. Avantek, which is now owned by Hewlett Packard, developed a series of M.M.I.C.s for use on relatively low frequencies. This range of amplifiers has a variety of specifications, with gains of up to 20dB and frequencies up to 2GHz.

For many r.f. designers these are ideal building blocks to use in a variety of situations. They are very easy to use, requiring input and output capacitors and a series feed resistor to limit the supply current as shown in Fig. 1.

They are never known to "take off" and give spurious oscillations, and provided that a few fundamental r.f. design layout precautions are observed they give excellent performance. They are also very cheap, costing only a few pounds. As a result they are often the answer to an r.f. designer's prayers.

These circuits are comparatively simple, containing a transistor amplifier but no strip lines or inductors. They are fabricated

The very much higher frequency circuits are often far more complicated, and sometimes consisting of a number of M.M.I.C.s placed onto the same substrate or carrier plate. When very high frequencies are required they are usually made from gallium arsenide.

# **Power Play**

Often high power M.M.I.C.s are required, especially for applications in transmitters. Here a new substrate material is

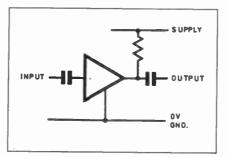


Fig. 1. Circuit diagram for an r.f. amplifier using an MSA series M.M.I.C.

helping to reduce costs and improve performance.

High power M.M.I.C.s experience problems with the efficient removal of the heat generated in the circuit. To overcome this, additional stages are usually required in the fabrication of the i.c. so that the correct heat removal techniques can be installed

This adds a considerable amount of cost to the final item. However, by using a new material called Silvar, developed by their Metallurgical Materials Division, Texas Instruments claim they can reduce the cost of a power devices by as much as 50 per cent.

tion to this a device having a low yield in manufacture is also less likely to have good long term reliability.

# Silvar Lining

Texas Instruments decided that the production of these devices could be greatly improved if a new concept in production was adopted. In a standard assembly the semiconductor itself is placed on a shim and then onto the substrate. Solder joints then connect the gallium arsenide chip to its external circuits, which might typically be thin film networks.

With the introduction of Silvar, a much simplified process can be used, removing much of the assembly time. The Silvar itself can be very accurately machined and this enables much less complicated assembly jigs to be used.

Another advantage of Silvar is that wire lengths can be shortened because circuits in the chip can be placed closer together. This considerably improves performance allowing the full potential of the chips to be realised.

The new material also exhibits the right co-efficient of expansion. It matches the gallium arsenide very well. In fact after all

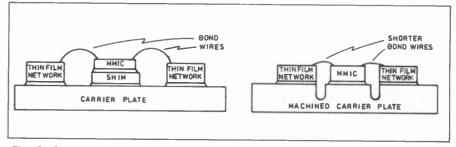


Fig. 2. Comparison of (left) existing M.M.I.C. method, and (right) new method using Silvar carrier plate.

In the production of specialised M.M.I.C.s and their assemblies there are three main areas which contribute to the costs. The first is the very specialised nature of the assembly, which means that automated manufacturing techniques cannot be used.

This calls for the use of very accurate assembly jigs which are very costly to produce. Highly skilled manual labour also has to be employed and this adds to the costs as well.

The final reason for the high production costs is that dimensional changes caused by temperature cycling which occur during manufacture can greatly reduce the yield. In order to ensure the production of any device remains viable, the number of rejects must be reduced to a minimum. In addithe processing stages it leaves the gallium arsenide chip in a state of slight compression. This has been proved to enhance the reliability of the chip.

The material is made from a combination of nickel iron alloys which are balanced to give the correct co-efficient of expansion. Some silver is also included to enable the high degree of thermal conductivity to be achieved.

Silvar is also very easy to machine, proving to be very much faster and hence cheaper than the existing materials which are used. Once the substrate is complete, the remaining stresses within the material are said to be very much lower than other materials. This is another factor which helps improve the reliability of the completed assemblies.

Is your PCB design package not quite as "professional" as you thought? Substantial trade-in discounts still available.

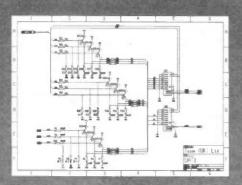
# **Board Capture**

# Schematic Capture Design Tool

- Direct netlist link to BoardMaker2

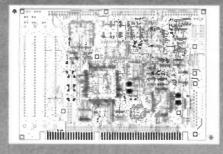
- Forward annotation with part values
  Full undo/redo facility (50 operations)
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# Board Maker

BoardMaker1 - Entry level

- PCB and schematic drafting
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  Full Design Rule Checking
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# PERSONAL PRACTICE AMPLIFIER



**ANDY FLIND** 

Headphone power keeps the family cordial while you strive for chords harmonial.

IKE many of the author's projects, the idea for this one began with a request from a friend. A lead guitarist with a heavy metal and blues band, he spends a lot of his spare time practising and always likes his music loud.

His wife and neighbours, needless to say, are not always quite so keen. In addition, he spends a certain amount of time "on the road", and for this a personal amplifier that could be used whilst travelling or in hotel rooms was considered a most desirable accessory.

# MIXED RESPONSE

These days, musicians (rich ones, anyway!) can purchase magic boxes full of microprocessor driven electronics that will generate just about any effect desired. Guitarist friend wished to combine the headphone output from one of these, a stereo source despite the guitar's output being mono, with the output from a Walkman playing taped music with which he wished to practise.

The initial request, therefore, was to produce a box containing a simple mixer for combining headphone level signals, and having an output capable of driving eight-ohm enclosed stereo headphones at fairly high volume. These are apparently preferred by musicians to the smaller 32-ohm types generally used with Walkman players as they provide greater volume and exclude external noise more efficiently.

Additionally, it was felt that a mono input accepting signals directly from the guitar should be provided. It was decided at the outset not to provide any tone or input level controls as all the signal sources to be used with the unit would have their own. Restricting the controls of the unit to a single volume control and an on-off switch would make it simple and compact.

# **OP.AMP MIXERS**

For readers not familiar with op.amp mixer circuits, it is probably worth taking a few moments to explain their action. The basic circuit, which is configured as a simple op.amp inverter, is shown in Fig. 1.

Using negative feedback through resistor R3, the op.amp strives to maintain a voltage at its inverting (-) input identical to the one applied to its non-inverting (+) input, in this case zero volts or "ground".

If input currents flow through resistors R1 and R2 the op.amp will try to maintain this potential by adjusting its output voltage to sink the currents through R3. For this reason the inverting input of this circuit is sometimes referred to as a "virtual earth" point.

The advantage of this for mixing applications is that because the input and output feedback resistors always have a "ground" potential at one end, the input currents are always directly proportional to the input voltages. Consequently, the output voltage is always directly proportional to the (inverted) sum of the input voltages since this

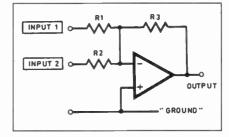


Fig. 1. Simplified op.amp mixer circuit.

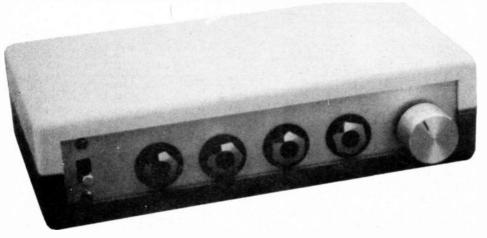
is what has to be applied to resistor R3 in order to maintain the "ground" potential. This applies where the resistors all have the same value, of course, so preserving unity gain.

With suitable adjustment to the ratio of input and output resistor values the circuit can also provide gain or attenuation. Additionally, since the inputs are always connected effectively to a fixed potential, they do not interfere with each other, an advantage in some applications.

# AMPLIFIER CIRCUIT

The full circuit for the Personal Practice Amplifier is shown in Fig. 2. Since it is a stereo circuit there are two mixers, one each for the left and right channels. The mixers are formed around ICla and IClb, the two op.amps within an LM358 device.

The four audio signals (two lefts and two rights) from the headphone sockets of the



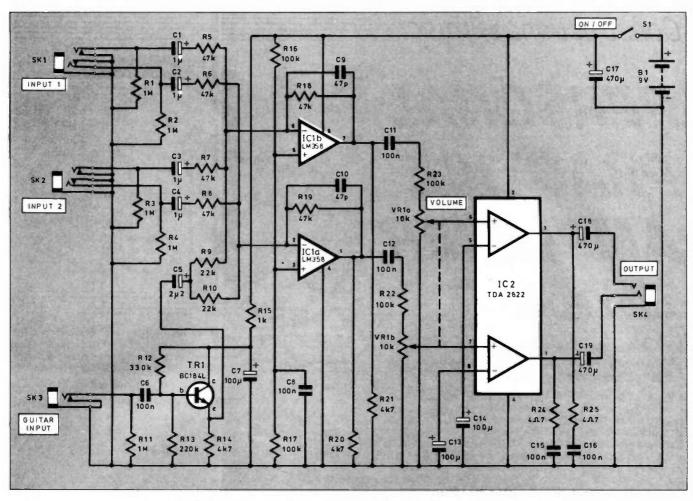


Fig. 2. Complete circuit diagram for the Personal Practice Amplifier.

two main music sources are a.c. coupled from stereo jack sockets SK1 and SK2 through capacitors C1 to C4 into resistors R5 to R8. The jack sockets are switched types, arranged so that unused inputs are connected to ground to eliminate any noise that might occur if they were left opencircuit.

High value resistors R1 to R4 are included to ensure that the input capacitors, and possibly the output capacitors of connected equipment, receive a "polarizing" voltage.

The signal from the guitar is input via mono jack socket SK3 and is buffered by transistor TR1 to provide an input impedance of around 120k, as required by the magnetic pickups of most electric guitars. This stage does not provide any voltage gain, the slightly higher gain required is achieved by the use of lower value input resistors R9 and R10 to the mixers. Both mixer stages are driven simultaneously by this input. The power supply to the transistor is decoupled by resistor R15 and capacitor C7 to ensure stability.

In the simple example of Fig. 1, it was assumed that the op.amp was supplied from a dual supply and the inputs worked at "earth" potential. In Fig. 2 it will be seen that there is only a single 9V supply, so the reference for the two mixers is set to half this by resistors R16 and R17, with a decoupling capacitor C8.

With two other minor additions, the mixers are virtually as described in Fig. 1. Since we are surrounded by assorted radio transmitters, from CB to portable telephones, it's always a good idea to define the upper frequency response of audio circuits. Capacitors C9 and C10 restrict the

upper limit to around 40kHz. The low frequency response of this circuit is around 20Hz, so bass players should find it just as useful as lead guitarists.

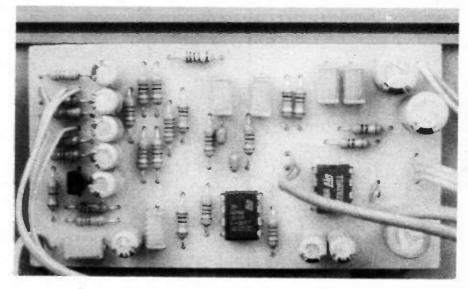
Next, the two op.amp outputs are slightly loaded by resistors R20 and R21 to negative supply, the reason for which may not at first be obvious. The LM358/324 op.amp family is not really intended for audio applications, and its Class B output stage can introduce unpleasant "crossover" distortion.

However, as these op.amps are cheap and simple to use, they are often employed in audio applications such as graphic equalisers without apparent problems. This is because, where low-power signal process-

ing is required, the performance can be improved with a loading resistor, as in this circuit. The small d.c. current taken from the output ensures that only one of the two internal output transistors ever conducts, effectively converting it to Class A operation and eliminating the cross-over problem.

# OUTPUT AMPLIFIER

From the mixers, the left and right signals pass through the volume controls VR1a and VR1b to the output amplifier. This is a TDA2822 stereo device, with two internal amplifiers each capable of delivering around 0.5W into an 8-ohm load. They have a fixed gain of around 40dB, and



the input signals from the mixers are attenuated by resistors R22 and R23 to compensate for this.

Outputs are taken to the output jack socket SK4 through the coupling capacitors C18 and C19. "Zobel" networks socket SK4 through R24/C15 and R25/C16, are provided in accordance with the manufacturers recommendations for stability. (Zobel is renowned for his work on filter designs.)

Power for the circuit is supplied by a 9V d.c. source. As the quiescent drain is only about 12mA, and is not much more than this at most normal listening levels, a PP3 battery is suitable. This should be an alkaline type as these cope more effectively with occasional peaks of current demand. The circuit continues to operate quite happily down to below 6V, so PP3 battery life should be adequate. Where a great deal of use is foreseen, a pack of six AA batteries could be used instead.

# CONSTRUCTION

The circuit is built on a single printed circuit board (p.c.b.) which is available from the EPE PCB Service, code 950. The layout

# COMPONENTS

Resistors R1 to R4,		
R11	1M (5 off)	
R5 to R8.	,	
R18, R19	47k (6 off)	
R9, R10	22k (2 off)	See
R12	330k	
R13	220k	SHOP
R14, R20,		TALK
R21	4k7	
R15	1k	Page
R16, R17,		
R22, R23	100k (4 off)	
R24, R25	$4\Omega7$ (2 off)	

Potentiometer

VR1 10k dual carbon rotary, log

Capacitors C1 to C4 1μ radial elect. 63V C5 2µ2 radial elect. 63V C6, C8, C11, C12 C15, C16 100n polyester (6 off) C7, C13, C14 100μ radial elect. 10V (3 off) C9, C10 47p ceramic (2 off) C17 to C19 470μ radial elect. 16V

Semiconductors

BC184L npn silicon TR1 transistor LM358 dual op.amp IC2 TDA2822 stereo power amplifier

(3 off)

Miscellaneous

SK1, SK2, SK4 stereo jack socket, switched (see text)

(3 off) SK3 mono jack socket, switched (see text) s.p.s.t. slide switch

Printed circuit board, available from the EPE PCB Service, code 950; plastic case 153mm x 84mm x 39.5mm; knob; battery connector (see text); 8-pin d.i.l. sockets (2 off); connecting wire; solder, etc.

Approx cost guidance only

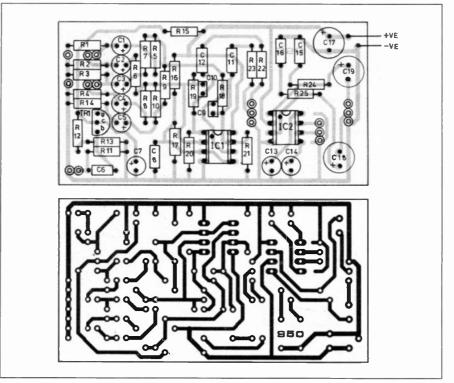


Fig. 3. Component layout and full size copper foil track master for the printed circuit

positions for all components are shown in

There should be no problems with constructing this circuit. The components could be simply fitted to the board, checked and powered up, but for the more cautious it can built and tested in stages if preferred.

All the passive components - resistors and capacitors can be fitted to the board first. Care should be taken to ensure that the electrolytics are fitted the right way round. Omit transistor TRI and the two i.c.s for the moment, although their d.i.l sockets may be soldered in. After thoroughly checking that your soldering of all the components is satisfactory, power can then be applied.

Apart from the charging surge and any minute leakage of capacitors C7 and C17, the current taken should be negligible. Anything more than a couple of hundred microamps indicates a problem and should be investigated before proceeding.

Next, transistor TR1 can be fitted, again taking care with polarity. This will add just over half a milliamp to the current drain, and the potential at its emitter should be a little over 2V. Now insert IC1, which will take the total supply current to about 3.5mA. The potential at ICI's outputs, pins I and 7, should be half the supply, or about 4.5V

If an oscilloscope is available, it is possible to monitor these outputs whilst injecting input signals to check operation of the mixer stages, but this is probably being over-cautious!

Before fitting IC2 it is necessary to connect volume control VR1a and VR1b and provide for monitoring the output,

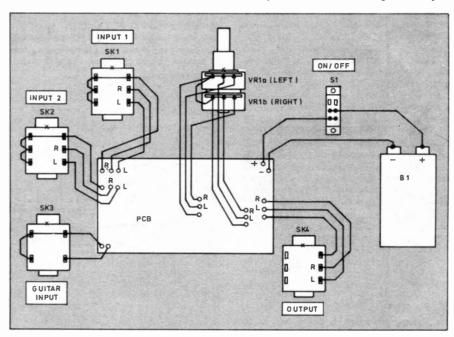


Fig. 4. Interwiring details for the Personal Practice Amplifier.

either with headphones or a 'scope. The fitting of IC2 will raise the total supply current to about twelve milliamps. The audio output should respond with loud hums when inputs are touched with a finger, the level depending upon the volume control setting.

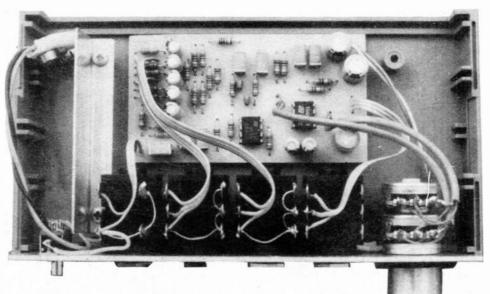
# CASE AND WIRING

Any type of case preferred by individual constructors can be used for this project, though the one suggested results in a very attractive, pocket-sized unit.

If a metal case is used, it should be connected to the negative supply rail, as should any metal panels in plastic cases. In the prototype this was achieved simply by soldering a short lead between the negative wiring on the nearest jack socket and one of the securing screws of switch S1.

The wiring connections shown in Fig. 4 are for the standard 0.25-inch switched jack sockets used by most musical equipment, though other types, such as 2.5mm or 3.5mm, can be used if preferred. Provided the input connection leads are fairly short, there is no need to use screened leads as the inputs are mostly of relatively low impedance and high signal level.

Volume control connections, however, should be made with screened lead, as these carry low level signal voltages at medium impedance and are therefore more sensitive to hum, etc. Connections to socket SK3 might need shielding if interference problems are experienced in use, though this was not the case with the prototype. Note the crossover of the signal connections to SK4, necessary to maintain correct "left" and right" signal paths.



A switched socket may be fitted for an external mains driven power supply. If this is done, it may prove necessary to earth the negative side of the supply to prevent mains hum, and to ensure safety.

# **AMPLE USES**

The output of this amplifier when driving headphones should certainly prove sufficient to satisfy the demands of most musicians. It may also find other uses beyond the application for which it was designed. If a large power supply is used, the outputs could supply a couple of

8-ohm loudspeakers, so it could be used as a mini PA or perhaps as a booster for a Walkman.

Another use would be as a general purpose workbench amplifier for experimenters. One note of caution: the volume should not be turned up so far as to drive the output into clipping, as under these conditions the output chip suddenly gets very thirsty, which translates as expensive in terms of batteries. It is unlikely, though, that many users would wish to use it at anywhere near this level for very long.

# SHOP TALK with David Barrington

Versatile Microcontrolled 3-Digit Timer

The little pressbutton switches used in the Versatile Microcontrolled 3-Digit Timer project may prove difficult to purchase from your usual local source. These are variously listed as "miniature p.c.b. keyboard keyswitches" or more commonly as "miniature tactile switches".

The horizontal version is used here and should be stocked by most component advertisers. However, in case of difficult similar switches are listed by Maplin as Tactile Switches, code KR92A and by Electromail ( 01536 204555) under Keyboard Switches, code 320-871.

The 0-5in. 7-segment displays should be available generally. Display XI *must* be the "high brightness" type. Also, to fit on the p.c.b. they must be the type with a "horizontal" pinout arrangement, i.e. pins run across the top and bottom of the device. Remember to specify a common anode and *two* common cathode devices when ordering

mon cathode devices when ordering.

Checking for supplies of the 3·2768MHz crystal, we were surprised how few component stockists carried a good selection of crystals. It is currently listed in the new Cirkit catalogue, stock code 45-03000. We only found it listed by Electromail (657-527) by looking at the end of their semiconductor section; they do not carry a "crystal" listing in their main Index.

Most of our component advertisers will be able to supply a suitable 12V relay. However, before ordering check that it will fit on the p.c.b. or be prepared to mount it off-board and hard wire it to the p.c.b. The Maplin 3 Amp Min Relay (YX96E) should sit directly

on the p.c.b. It is most important that whichever relay is used the CONTACTS are suitably rated for switching the attached appliance.

A pre-programmed PIC16C54 microcontroller is available (strictly mail order only) from the author, price £9.50 including postage. All monies should be made payable to B. Trepak. Orders should be sent to Mr. B. Trepak, 20 The Avenue, London, W13 8PH. Overseas customers add £2 to order.

The Timer printed circuit board is available from the *EPE PCB Service*, code 933.

# Solar Seeker

The choice of motor and transmission caused the authors the greatest number of headaches when selecting parts for the *Solar Seeker*. After a lot of head scratching a radio control (RC) servo motor was finally adopted here.

The motor in the prototype model is designated "Acoms IC AS-7 Servo." We understand from the designers that most model shops should be able to offer a suitable RC servo motor, with gearbox. They also pointed out that "RC motor with mounting hardware" is also referred to by some model makers as an "RC Servo horn" and is a small plastic coupling stocked by RC model shops.

They also took some time trying to assess the leakage current in tantalum capacitors and found that the Maplin sourced ones were better than some others tried – Maybe a statistical freak! Anyway, don't forget that capacitor C1 must be a tantalum type.

The reed relay is a sub-miniature type rated at 5V d.c. 10mA, with a 500 ohm

coil. A suitable type is listed by **Maplin** code JH12N. The dual power op.amp type L272M is also listed, code UJ36P.

# Infra-Red Remote Control Unit

The only item that could cause any real concern when undertaking the *Infra-Red Remote Control Unit* is likely to be the UM3750 encoder/decoder i.c. used in both the Transmitter and Receiver circuits. If any problems do arise, it can be ordered from Maplin, code UK77J.

The red plastic lens, used over the Receiver photo diode, also came from the same source, code FA95D. The TBA2800 infra-red preamplifier i.c. is currently listed by Maplin (JU36P) and Cirkit stock number 61-02800. Most of our component advertisers should be able to supply the TIL100 infra-red photo diode or its equivalent.

The only other part called up for the Transmitter that needs careful selection is the infra-red emitting diode. This must be a "high power" type, such as the TIL38.

The 6-way d.i.l. switch is now a stock line with most of our advertisers and should not be a problem locating. The Transmitter and Receiver printed circuit boards are available from the *EPE PCB Service*, codes 948 (Trans.) and 949 (Rec.) respectively.

Ramp Generator

Some of the items called up for the Ramp Generator need to be specially purchased if the constructor is to adhere to the published design. Information on these items was given in Shop Talk last month.

# **Personal Practice Amplifier**

Not too much can go wrong when ordering parts for the *Personal Practice Amplifier* project. Remember to quote the suffix L when ordering the BC184L transistor and to ask for a stereo dual ganged "log" potentiometer.

The TDA2822 stereo power amplifier i.c. was purchased from Maplin, code UJ38R.



# **ID Cards**

The UK government will soon decide whether to introduce an electronic identity card scheme. The main options are to do nothing, make cards available on a voluntary basis, treat a driving licence with photo as an ID card, or make it compulsory for everyone to carry a government card which stores full personal details. Everyone will have their own views and anyone can send their comments to the Home Office before 30 September.

In a "special feature" (*EPE* July '95) I described how Smart Cards could be used to store a wide range of picture and text information, but at a relatively high cost. This is one argument against a compulsory scheme. But two recent developments have moved the goalposts.

US giant Motorola has joined with Matsushita of Japan and bought the Indala Corporation, a specialist electronics company in San Jose, to offer new memory technology which cuts the cost of memory storage. Kodak and IBM have developed a "hypercompression" system that stores photographs in such a small memory space that even a magnetic credit card can carry the owner's electronic picture.

# Radio Honeycomb

Existing memory cards either have a built-in battery to hold data inside volatile RAM, or use high currents to write to non-volatile memory EEPROM (Electrically Eraseable Programmable Read Only Memory). A new material, Y1 is a ferroelectric ceramics material similar to a high temperature superconductor and when formed into a honeycomb lattice of individual cells, the cells store bits of digital code as isolated spots of capacitive charge. The written charge pattern remains permanent until a fresh writing current is applied.

The Y1 theory was proposed five years ago by the University of Colorado, but Matsushita of Osaka developed and patented a way of making a cell matrix by a technique similar to that used to fabricate microchips from silicon wafers. Indala then worked out a way of burying the matrix inside a credit card or resin button, along with a miniature radio transponder and aerial made from 400 turns of very thin wire. The matrix uses the power of an interrogation signal to send back a response signal which is coded to carry whatever information is in the memory.

In 1993 Motorola bought Indala and also signed a joint venture deal with

Matsushita. Motorola and Indala design the Y1 cell structures which Matsushita fabricates in Osaka.

"It is very nearly the perfect memory material" says Rudyard Istvan, the Motorola President of Indala. "But if you do not use exactly the right recipe for the mix, the cake turns out flat".

# **Energy Levels**

Structures of Y1 need less then 1V, and energy levels of only 20 picojoules to read and write to the cells. This compares with the 12 volts and 200 picojoules needed to write to EEPROM or Flash memory chips used in smart cards.

Whereas Y1 reads and writes at the near instantaneous speed of RAM, Flash and EEPROM takes one second to read each kilobit of data. Also Flash and EEPROM memories start to fail after they have been written to and erased a few hundred thousand times. Y1 survives a billion cycles.

For identity cards the reader emits a signal at 125kHz, from a distance of up to one metre. For car tolling, the reader transmits at 2·4GHz, and can reach over a range of 100 metres.

The only limiting factor is the minimum working size of the Y1 cells. These cannot be less than one micrometre across, and this limits the practical capacity of a Y1 memory chip to 256 kilobits or 32 kilobytes, or around 3000 words of text. Although small by today's PC standards this is more than enough for identity cards.

Istavan says Motorola can drive the cost of cards down to \$2 each, which is less than half the cost of a smart card with equivalent memory, and none of the practical limitations. Tag buttons will cost \$1. Readers can be made for \$75.

# **Image Bank**

Says Istvan "We have already thought of using the system to store electronically compressed photographs. You could walk through passport control without taking the card out of your pocket. The only way to conterfeit cards would be to build a silicon wafer fabrication plant and modify it for Y1 processing".

Kodak's Image Verificaton System (IVS) works with a reader developed by IBM to display an image of the cardholder's face on a TV screen. The display terminal, says Kodak, can either be a cash register or an identity verification terminal.

"We are proposing IVS as an industry standard", said Carl Gustin, Vice-President

of Kodak's Digital and Applied Imaging Division.

A full colour image normally requires at least a megabyte of storage space, and fills a floppy disk. Existing compression technology can reduce this to a few tens of kilobytes, without too much quality loss. But this is still far too much data for a card with affordable memory capacity.

To be useful as an identification tool, the card must also store text which needs one byte for each character. The magnetic strip on an ordinary credit card has a storage of capacity of under 250 bytes.

By using IVS it squeezes the image down to an unprecedented 400 bits, or 50 bytes. This leaves plenty of space on even the cheapest one kilobyte smart card for personal and medical text. It is also small enough to store on a magnetic credit card.

# On Track for Photo-fit Identity

Kodak suggests putting the picture in the third of the three magnetic tracks which are defined by the ISO card standard. According to this standard two tracks contain information about the holder and can only be read by a card reader.

The third track can be written to by a reader as well as read. This third track has a capacity of 107 bytes and is usually left half empty by the banks. So a standard credit card can store a 50 byte hypercompressed picture without limiting other functions.

Although Kodak will not explain how the technology works, the black and white pictures displayed by IBM's reader suggest that no attempt is made to store a true photographic image of the holder. Instead the 50 bytes convey flag signals which trigger the reader into searching through a pre-programmed memory of facial characteristics.

The reader then blends these together into a photo-fit image. Dithering, the addition of a little random noise, smooths the joins between the component parts of the image. Because so little information is needed, the same technique can be used to store pictures on bank cheques as a bar code string.

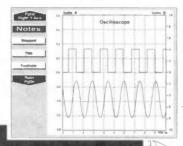
Although the images so far displayed are only in black and white, Kodak claims that colour recording of passport picture quality is also possible if the store capacity is increased to three or four kilobytes.

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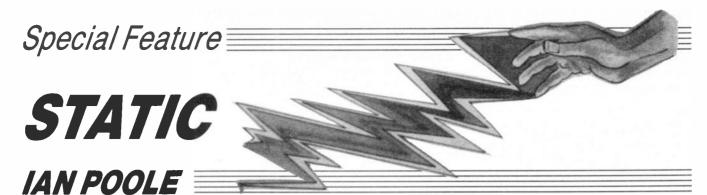
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Understand the nature of static electricity and how to avoid its problems in electronics.

TATIC is a problem which is becoming increasingly important in electronics. In the very early days of valves, no one gave it a thought. When transistors arrived, it was still not a real problem. However, when MOSFETs arrived they started to fail for no apparent reason as people began to discover that static could have disastrous effects on semiconductors.

Nowadays, manufacturers of electronic equipment spend vast amounts of money investing in equipment to ensure that semiconductor devices are not exposed to static or as it is more correctly termed electro-static discharge (ESD). Indeed many manufacturers treat all semiconductors as static sensitive devices, not just the MOS devices which are most prone to damage.

# WHAT IS STATIC?

Static is an everyday fact of life. It is simply the build up of a static electrical charge between two bodies. From time to time this build up is discharged. This can take place in a very short space of time, and there can be some surprisingly high levels of current and voltage.

Lightning is, of course, the most dramatic result of static discharge. Potentials of many thousands of millions of volts are reached and the currents can be as high as several tens of thousands of amps. This is all discharged in a matter of a few milliseconds. With these levels it is hardly surprising that it can cause serious amounts of damage when it occurs.

Even in the electronics workshop, in the office, or around the home very high levels of static can build up. It is not unusual for someone to build up a static charge of ten or even twenty kilovolts (kV) simply by moving around in the home.

The discharge of this type of static is much faster than that of lightning. It lasts a matter of a few tens of nanoseconds and the peak currents can be as high as several tens of amps. It is hardly surprising that occasionally one can feel a tingle as the discharge takes place!

Obviously the levels of voltage and current which are produced depend of a large variety of factors. Everything from the size of the person, to clothing being worn, the object against which the discharge is made, and of course the humidity of the air. These all have a pronounced effect so it is almost impossible to predict the exact size of the discharges which are possible.

# HOW IT AFFECTS ELECTRONICS

Problems caused by ESD came to the fore with the arrival of metal oxide silicon (MOS) technology. When these devices were introduced many had input resistances well in excess of a thousand megohms.

Whilst this seemed the answer to many design engineers' dreams it also spelt disaster as these devices were only able to withstand voltages between the gate and drain of about 40 volts maximum. Even people walking nearby could damage them if they were carrying a large charge!

Responding to this problem, manufacturers integrated protection diodes into their devices. Even so they often put a shorting protection wire around the leads which was to be removed only when the devices were safely installed in their circuits. Once in their circuits they were relatively safe because there would be sufficient leakage paths around the device to prevent damage under normal circumstances.

Many investigations were made into the effects of ESD in view of the enormous possible costs to industry. It was soon discovered that even though many devices failed immediately after they had been subject to a static discharge, others would carry on working for a while, or they would have their performance reduced.

Many studies were undertaken by a variety of organisations. Obviously those involved in space research were very keen to ensure the maximum reliability of their equipment. NASA undertook some large investigations to determine the long term effects of static.



Typical anti-static plastic case and conductive foam. May be used with larger or more expensive i.c.s.

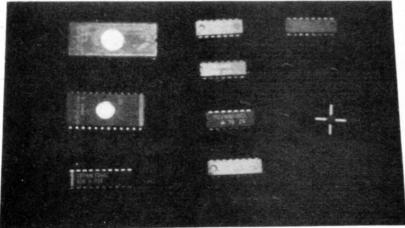
They analyzed the results of the electronic failures that had occurred during the *Apollo* programme. As a result they now attribute over 50 per cent of the failures to the effects of ESD, both in the short and long term.

Originally only MOS devices were treated as being static sensitive. Most electronic enthusiasts who have used the original 4000 series of CMOS chips will remember the extreme care which had to be used when handling them. Radio amateurs will also remember the problems when handling discrete MOSFETs which were very popular for v.h.f. preamplifiers.

These devices can be damaged by discharges of static voltages of between 100V and 200V. However, the devices which must take the prize for sensitivity to ESD must be gallium arsenide f.e.t.s. They can be destroyed by discharges as small as 5V.

Nowadays, virtually all semiconductor devices should be treated as static sensitive. A typical discrete bipolar transistor can withstand the sort of discharge which can just be felt by the human hand.

However, as the size of components within integrated circuits shrinks it is found



Anti-static conductive foam should be used for storing loose i.c.s. It is available in large sheets from many electronics stockists.

that bipolar i.e.s are becoming just as susceptible as their MOS counterparts.

In fact even some passive devices like capacitors and resistors can be damaged in some instances. With the ever decreasing size of components with the surface mount technology revolution, even those devices which were once thought safe from the effects of ESD are found not to be immune.

In the future, matters are likely to become even more critical. With feature sizes on i.c.s shrinking even further, as more is added to each chip, they will become even more susceptible. The addition of more components to each chip will mean that they become more expensive, making the results of static even more costly.

# MODES OF FAILURE

In looking at how ESD affects semiconductor devices it is necessary to remember that it is not just the voltage that is important. The time taken for the discharge must also be taken into consideration. Often the discharge takes around 100ns or less.

The actual time depends on many factors, but it is found that discharges from sharp points, e.g. the end of a small screwdriver, or tweezers are much faster than from a relatively blunt object like a finger. There can be a factor of five or more difference! This means that if the same amount of energy is discharged the currents can vary by five or more, being higher when sharp instruments are involved.

It is quite possible that a person charged up to 20kV can give a peak current of 20A. When it is considered in these terms it is easy to see how the possibility of damage arises in i.c.s.

When compared with the size of some of the elements in the semiconductors these currents are colossal. In fact even relatively small discharges can produce significant amounts of heat.

Even though the discharges are very short lived, the architecture is so small that the heat cannot be dissipated in time, and elements can be melted or even vaporized. Often semiconductor junctions are the victims, and it is found that the time needed to destroy a junction is inversely proportional to the duration of the pulse.

When a semiconductor junction breaks down, the high voltages present cause it to reverse avalanche. This happens as the electrons are accelerated by the electric fields associated with the high potential difference across the junction.

As these electrons move faster they collide with other atoms releasing further electrons which in turn release more. With the high currents which are available from the ESD sufficient current can pass through the junction to cause it to overheat and melt. This will leave a low impedance short circuit across the junction rendering it useless.

Other failures can occur when a track joining areas of the device is vaporized leaving it open circuit.

Not all discharges result in immediate failures. The aspect of more concern is that the device can be left with a long term weakness which will fail later in its life, or it can result in an intermittent failure.

This occurs when insufficient damage is done by the discharge to completely destroy the track or junction. A track may have its width reduced, and as it may no longer have sufficient width to carry the required current it can fail when even small overloads are placed on it during normal operation.

# COMBATTING THE PROBLEM

In view of the problems with ESD, it is obvious that even the home constructor must take precautions to avoid its effects.

Whilst it is not possible for the amateur enthusiast to take the same steps as commercial manufacturers there are a number which are quite easy to implement, and will avoid most of the problems.

When looking at how to combat ESD it is necessary to look at all stages of the life of the i.c.. This means that just as much care has to be taken when handling and storing the chip as it does when soldering the chip into the circuit board, and then handling the completed circuit board.

To help reduce the possibility of ESD damage a wide variety of products are available. Firstly, conductive foam should be used for storing loose i.c.s. The chips can be easily inserted into the foam, thereby ensuring that all the connections are shorted together.

However, before using the foam it is best to make sure that it is conductive. Although it is often black in colour, not all black foam is conductive. A simple test by applying the two probes of a multimeter switched to the resistance range will reveal whether it is conductive. If it is then the meter should read a value of a few hundred ohms at the most.

This foam is now available from many electronic stockists. Whilst it may appear a little on the expensive side to buy, a single sheet is more than sufficient. In fact it may be worth buying a sheet with one or more friends.

Anti-static i.c. tubes are also available. Often when buying a few chips, they may arrive in these tubes. If they do, then keep the tubes for use in the future.

For handling completed circuit boards special anti-static bags are available. Often it is not worth buying these bags at a stockist. However, many ready built boards now come in anti-static bags. Computer boards are a very good example and it is worth keeping these bags when the opportunity presents itself.



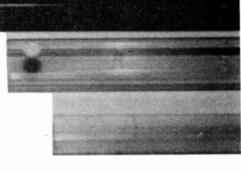
Anti-static bags are frequently used to store i.c.s, complete board assemblies and other static-prone components.

# **CONSTRUCTION**

Precautions are also very important when fitting static sensitive devices into circuits. A number of steps can be taken here to ensure that the risks of damage are reduced.

First of all it is necessary to make sure that you are not charged up. This can be done by ensuring that you are earthed.

Special wrist straps are used by most electronics companies to ensure that people are not carrying any charge. The strap itself consists of an elasticated band containing conductive fibres. A lead with a resistance of one Megohm or more is then used to connect the strap to ground. This resistance is quite sufficient to ensure that any charge which is built up is safely routed to earth.



If your i.c.s arrive in anti-static tubes, keep the tubes for future use.

Static dissipative work benches can also be installed. New work stations being installed generally have these surfaces. They should obviously not be totally conductive otherwise they will short out any circuits placed on them. Again they have sufficient conductivity to remove any charge in a controlled fashion.

Another alternative is to use semi-conductive rubber mats. These are popular because they are quite cheap and are an ideal solution for the home experimenter. Some stockists sell all the parts for a complete anti-static work station for as little as about £25. Although this may appear to be a significant amount of money, it may prevent a project from damage and being consigned to the cupboard to gather dust because it does not work.

The anti-static mats and wrist straps are normally connected to earth, usually via a large value resistor. As mains earth is normally used, extreme care must be taken to ensure that the work station does not become live, even though the resistors will help to reduce the danger. Special yellow plugs with only an earth connection are available for this, and can help reduce the danger. Even so, check that the system is totally safe before using it.

Care should also be taken when using tools. Most items with metal handles are perfectly fine. However, many have insulated handles and if these have to be used they should be discharged before use and kept on the anti-static area.

The most important tool is the soldering iron. These days it is possible to buy irons with earthed bits. These should be used at all times. Not only are they better from a safety point of view, but they also do not build up static.

# CIRCUIT DESIGN

Circuits should also be designed to minimise the effects of static. If not then even completed boards can be sensitive to the effects of static.

The major precaution is to provide a leakage path to earth for all lines which come onto a board or go off it. This is particularly important where logic families such as the ordinary 4000 series of chips or the more modern 74HC and 74HCT series of chips are concerned.

For these circuits it is usually just a matter of adding a pull-up or pull-down resistor of about 10 kilohms or 100 kilohms from the line to either supply rail. In this way any static build up can be prevented and any static which may be discharged onto the lines will have a leakage path to earth.

# SUMMARY

Industry is spending millions of pounds in protecting against the effects of static. Home constructors and enthusiasts also need to be aware of the effects of static on electronic devices.

Even though it is not feasible to go to the same lengths as manufacturing companies, there are many precautions which should be taken which do not cost vast sums of money. By doing this it is quite possible to save many hours in fault finding devices which have failed as the result of static.

Providing that the simple basic ESD precautions are observed, readers should not be unduly worried about using static-prone components.

# Table 1. Basic anti-static handling precautions.

Most of these rules are easy to apply and only require a little forethought. If in doubt, always treat a device as staticsensitive - it is better to be safe than sorry!

- Static sensitive components are normally supplied in special packaging which is designed to safely dissipate static charges. Components should not be removed from this packaging until they are required to be put into service. The packaging should then be retained for future use.
- Store components that are susceptible to static in their original packaging, using electrically conductive containers or conductive plastic foam.
- Avoid contact with the pin connections of a device whenever it is "out of circuit".
- Regularly discharge your body before handling components or circuits. You can very easily do this by simply touching a grounded point (such as the

metal case of an oscilloscope or power supply) with first one hand and then the other.

- Ensure that you have a "grounded area" in which to work. A rubber "earth mat" makes an ideal area in which to keep your devices and circuits while you work on them.
- Avoid the use of all static producing materials. In particular it is worth keeping polythene bags and other plastic materials well away from your work area.
- Make use of a low-voltage soldering iron and ensure that the metal parts of the iron are properly earthed. It is also worth checking that all of your other mains operated equipment is properly grounded.
- 8. If you have a computer or TV receiver in your workshop it is important to keep the display well away from static sensitive circuitry. Very high static voltages (usually well in excess of 10kV) are present in the vicinity of a cathode ray tube.
- Finally, if you regularly handle static sensitive devices, consider using an earthed wrist strap!

Table by courtesy of the *Electronics* Service Manual.

# Ohm Sweet Ohm Max Fidling the electrical outlet happened to be on the

# Fishy Goings-On

The domestic scene at Chez Fidling had very recently gained a new element of entertainment which appealed to Piddles, my feline friend, probably in more ways than one: I had bought a tropical aquarium. A small glass tank, now suitably replete with gurgling decorations (a tasteful plastic icon of a deep sea diver, to be precise), plastic seaweed, fluorescent lighting and a thermostat added to the general interest, whilst my home-made and freshly painted wrought iron fish-tank stand completed the scene. Of course, the fish themselves were just incidental, of much more interest was the wiring up and the general alchemy involved in successfully commissioning my new aquatic acquisition!

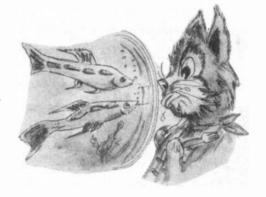
Justly proud of this latest project, I had switched it all on to let everything settle down for a week or so, before cycling over to the local pet shop to invest in some brightly-coloured guppies. These were successfully introduced and were swimming merrily away in the new tank. The puss entertained himself periodically by nudging the tank with his nose, in close proximity to my piscine pals inside. Needless to say, my attention soon turned to the electrical side of things and before long I had determined that I could "improve the performance" of the fish-tank with the addition of some electronic goodies. Preferably with lots of flashing light-emitting diodes, like all good projects have!

One major worry (such was the heavy burden of responsibility I carried as a newly-found fishkeeping expert) was to ensure that the fish-tank's pump was never accidentally switched off. The Boss would occasionally unplug it when doing the hoovering, much to my chagrin. However, the electrical outlet happened to be on the same circuit as my workshop's, so whenever I fused the wiring in the 'shop - which was pretty often - the fish-tank's pump also went a.w.o.l.; in the guppy scale of things, pretty catastrophic! Perhaps I could develop a gadget which would help avoid this perilous position ...?

# Under pressure

I retreated to the workshop to ponder a plan. I have a large cardboard box full of ex-equipment paraphernalia, mainly taken from old washing-machines, dishwashers, fridges and the like. I lifted the box from the shelf and rummaged amongst the water pumps and motors within. Entwined in copious lengths of quarter-inch rubber tubing was a water level switch that I'd removed from a defunct washing machine. This was a round device, roughly four inches in diameter with a spout. To my delight I had discovered that if I blew down the spout, a "click" would ensue from its innards. Intrigued by this phenomenon I tested out the switch contacts and discovered that what I was holding was a pressure switch with a set of changeover contacts. A rare find indeed!

Wiring this up to an audible warning device and 9V battery, I had invented the world's first automatic, electronic, fish pump failure alarm! The sounder was a fiendishly piercing piezo siren, deafening and rather excessive for this application but it was all I could find in my biscuit tins full of bits. The gadget was soon spliced into the pump's air supply tubing using some black rubber hose taken from the old washer. Although the corroded pressure switch looked rather incongruous against all the new pump equipment, it was effective and cheap and was pressed into immediate service. Mission accomplished, I



retired to the 'shop to carry on tinkering with my latest magazine project whilst Piddles sat transfixed with tail swishing, licking his lips and peering up at the tropical fish swimming inside.

In the workshop my latest constructional project was taking shape: a thermostat which I hoped to adapt for the fishtank. Bristling with thumbwheel switches, sockets and l.e.d.s (naturally), this looked impressive and having quickly wired the mains side up, I was ready for the switchon and the first test. This I did with the usual flourish, full of confidence in my abilities.

At precisely this point, there was a dull thud and blue flash from the plug, and everything electrical went off. I had fused it all again! Approximately five seconds later, Piddles was seen rocketing out through the cat flap in the kitchen door as he disappeared round the corner. Yes indeed - the fish-tank pump had also stopped working, and the alarm had sounded, petrifying the pesky puss who scooted out the door! They say revenge is sweet. In my case revenge is rare, therefore sweeter still! Upon his return I proffered some bread and honey as compensation which Piddles scoffed greedily (as usual), whilst eyeing the fish tank warily.

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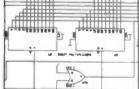
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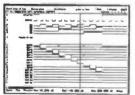
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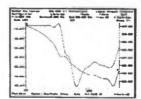


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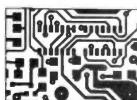
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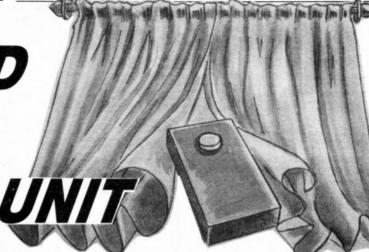
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INFRA-RED
REMOTE
CONTROL UNIT



MAX HORSEY

P.C.B. Design by James Cohen

Produces a coded infra-red signal, suitable for use in alarm systems, door openers, light controllers etc.

Designed particularly for last month's Automatic Curtain Winder project.

HIS article shows how an infra-red control system for on/off control of any other equipment can be designed, and is particularly tailored for the Automatic Curtain Winder project last month.

The essential design points are:

- 1. The system operates over a distance of several metres
- 2. The receiver can operate on the same 12V supply as the Curtain Winder
- 3. The receiver is not triggered by daylight, room lights, other remote control units etc.
- 4. The Transmitter is small and portable

### WHERE DO WE START?

One of the problems associated with infra-red control projects is that when they fail to work (like most circuits when first tested!) it is hard to know whether it is the Transmitter or Receiver at fault. However, most people now own commercially built remote control units and although these will not operate the decoder section of the receiver, they are helpful in testing the sensor and amplifier stages if the full circuit fails to work. We will begin with the infra-red transmitter and encoder.

### INFRA-RED TRANSMITTER

The principle by which the Transmitter operates is shown in the block diagram Fig. 1. The circuit is basically an encoder followed by a Darlington driver capable of delivering quite a large current to a set of three infra-red light emitting diodes (l.e.d.s).

Most people have at least one, and in many cases several remote control units, and coding/decoding is necessary in order to differentiate between them. Coding/decoding has the additional advantage that a circuit can be programmed to "look" for a particular signal even

though it may be much weaker than other

The UM3750 encoder/decoder i.c. is designed to send a particular code depending upon which of 12 input pins are connected to 0V or left open circuit. This provides a total of 4096 possible combinations. A further UM3750 in the Receiver can be set to the same combination and will provide an output when it receives the precise code which is transmitted.

infra-red sources such as daylight.

Note that the system is designed for a single transmitted signal; for example, to

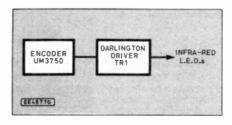


Fig. 1. Transmitter block diagram.

open a garage door, or close a pair of curtains. It *could* be adapted for a multichannel system, such as the type used to control a video recorder, but there are other specialised i.c.s available for this purpose.

The UM3750 can be used with radio control, ultrasonic control or infra red, and Fig. 2 shows how the encoding is accomplished. The pushbutton switch S13 must be pressed to transmit the encoded signal. Note that the other switches (S1 to S12) shown in Fig. 2 can be toggle switches, p.c.b. mounting d.i.l. switches, or simply wire links, inserted or not inserted to make a particular code.

When experimenting, all the pins 1 to 12 can be left open circuit (i.e. not connected); the only reason for connecting some to 0V is to ensure that another remote control unit (with the same circuit) will not operate your device accidentally. The combination of resistor R1 and capacitor C1 sets the frequency of the signal.

Depending upon the control and transmission system employed, R1 and/or C1 may need to be changed in value. However, it is essential to make identical changes in the receiver decoder circuit.

### TRANSMITTER CIRCUIT

The full transmitter circuit diagram is shown in Fig. 3, and consists of the encoder, plus a Darlington pair driver circuit, and a method of making the output into a short burst of power, to reduce battery consumption. The Darlington pair may be either a single npn Darlington transistor such as TIP121 or TIP122, or may be comprised of two separate npn transistors such as BC184L (or similar) followed by a power transistor such as TIP41A. The p.c.b. is designed for a single Darlington transistor.

Resistor R3, which limits the current through the infra-red l.e.d.s D2 to D4, is rather low in value and appears to allow a total of 1.8A to flow. Clearly neither the l.e.d.s nor the battery would be able to tolerate this current, so the circuit contains an additional 100 ohm resistor R4 which limits the total current to less than 100mA.

This would limit the possible operating range and so the smaller current is used to charge the 1000µF capacitor C4. When the pushswitch S3 is pressed a large

### INFRA-RED REMOTE CONTROL - TRANSMITTER

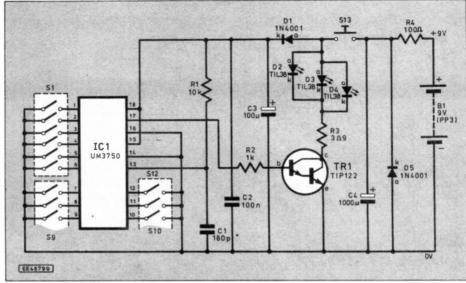


Fig. 3. Complete circuit diagram for the Infra-Red Remote Control Transmitter. Switches S1 to S12 are made up from two 6-way d.i.l. switches.

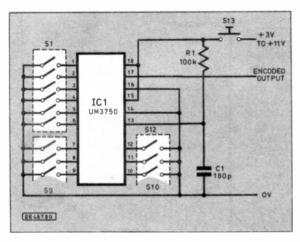


Fig. 2. Circuit diagram for the encoding stage of the transmitter. The code is set by selection of the switches S1 to S12.

current flows from the capacitor through the l.e.d.s for a short time. The pulse is too short to cause any damage, but provides the punch necessary to increase the range of the Remote Control System to several metres.

Diode D1, with the help of capacitor C3 maintains a 9V supply for the encoding circuit even whilst the voltage across C4 is collapsing. The diode D5 prevents disaster to the circuit if the supply is accidentally reversed. Such action will cause the battery to short circuit through the diode and resistor.

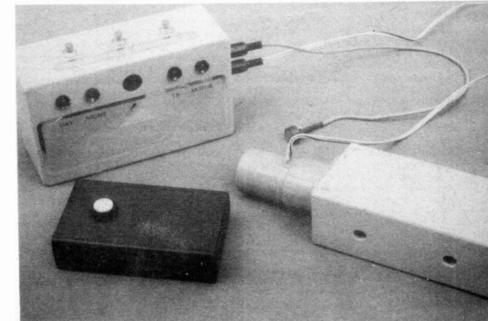
This is preferable to destroying the l.e.d.s, capacitor etc. A series diode could be used instead of the parallel diode shown, but its forward voltage drop may be unacceptable since every ounce of power is needed.

### RECEIVER CIRCUIT

The block diagram for the Receiver is given in Fig. 4. The Receiver requires several modules; first a high power amplifier followed by a decoder, programmed to look for the encoded signal from the Transmitter. The decoder triggers

COMPONENTS TRANSMITTER Resistors 309 10k 1000 R2 R4 All 0.25W 5% carbon film Capacitors 180p disc ceramic C1 C2 100n disc ceramic 100µ radial elect., 16V CA 1000μ axial elect. 16V Semiconductors D1, D5 1N4001 50V 1A rec. diode (2 off)
D2 to D4 TIL38 high power infra-red emitting diode or similar (3 off) 1N4001 50V 1A rec. diode (2 off) TIP122 n.p.n. Darlington transistor TR1 UM3750 encoder/decoder Miscellaneous S1 to S12 6-way d.i.l. switch (optional - see text) (2 off) S13 miniature push-to-make switch B1 9V battery (PP3), with clips
Printed circuit board available from the EPE PCB Service, code 948; handheld remote control plastic case, with battery compartment, size 103mm x 62mm x 23mm; 18-pin d.i.l. socket; multistrand connecting wire; solder terminal pins; solder etc. Approx cost guidance only

The completed Infra-Red Remote Control combined with last month's Curtain Winder project. The Receiver board is housed in the curtain control case.



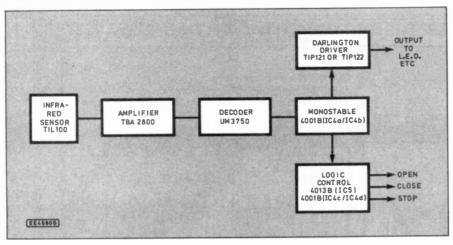


Fig. 4. Block diagram for the Infra-Red Receiver.

a monostable, which in turn lights an l.e.d. to provide the user with a visual indication that the message is being received.

The output from the monostable also triggers a logic circuit which translates the single messages into the required sequence of Open, Stop, Close, Stop to operate the Automatic Curtain Winder featured last month. Of course, the circuit may be used to operate any system; the l.e.d. signal switches on each time the Remote Control Transmitter button is pressed; the Stop signal switches on at every other press (i.e. on/off/on/off etc.) and the Open signal (alternating with the Close signal) switches on at every fourth press (i.e. off/on/off/off/on/off etc.)

The TBA2800 amplifier i.c. was chosen to provide the necessary gain, together with an infra-red photo diode similar to a TIL100. The signal available from an infra-red photo diode is rather small, and may contain unwanted interference from other sources such as lights, daylight etc. The amplifier is therefore carefully designed to reject this interference yet still provide a large amount of signal gain. The TBA2800 achieves this by means of three stages of amplification and includes automatic gain control (a.g.c.) – a very complex circuit in a small, low-cost package.

The complete circuit diagram for the Receiver is shown in Fig. 5, and the internal circuitry of the TBA2800 (IC1) is illustrated, in block diagram form, inside the dashed lines. The only external components required are three capacitors, C2, C3 and C4. The i.c. does the rest! Note that resistor R1 together with decoupling capacitor C1 provides a very stable power

supply for the i.c.

The pin numbers of IC1 are shown around the dashed lines. Two outputs are provided, a "normal" output from pin 7 and an "inverted" output from pin 8. The values of the capacitors have been chosen in relation to the signal being processed as discussed later. Some experimentation may be necessary if other signals are required.

IMPORTANT NOTE: IC1 must be used on a supply of 5V, not 9V or 12V. A 4·5V battery will do, or ideally a 5V regulated supply.

### DECODER

The signal from the output of IC1 (pin 8) is delivered to pin 16 of IC2, the decoder. The decoder is very similar to the encoder in the Transmitter.

Providing the same combination of pins (1 to 12) are tied to 0V as in the encoder, when the correct code is received pin 17 goes low for a short time (about 0.1s). In

fact, IC2 waits until it has received four valid codes in quick succession to ensure against false triggering. Many repeats of the code will be transmitted when the Transmitter switch is pressed and so there is no apparent delay whilst the four valid codes are checked.

The output from IC2 is capable of sinking 2mA when low (active). Although this could be used via a transistor circuit to operate an l.e.d., the short time for which the output pulses low suggests the need for a latching circuit or a monostable. A monostable allows a short pulse to trigger a pulse of any duration as required.

### **POWER SUPPLY**

We should at this stage pause and examine the power supply requirements. The amplifier IC1 requires 5V. IC2 will work on supplies of between 3V and 11V. A supply of 5V is therefore suitable for IC1 and IC2. However, the Curtain Winder module (last month) and hence the rest of the circuit requires 12V.

Regulator IC3 is, therefore, included in order to supply 5V for IC1 and IC2. Capacitors C6 and C7 provide decoupling, and diode D8 prevents damage to IC3 if the 12V supply is switched off and the voltage at the input side of IC3 becomes less than at its output due to the charge stored on capacitor C7.

# TRANSISTOR INTERFACE

The output from pin 17 of IC2 is normally "high" (about 5V in this case). This is connected via resistor R3 to the base of transistor TR1.

The current flowing through R3 keeps TR1 switched on. Hence the voltage at the collector (c) of TR1 is low (about 0V) and since this is connected to pin 1 of IC4a, it also will be low.

When a valid code is received pin 17 of IC2 will go low, turning off TR1 and causing its collector voltage to switch to 12V, thanks to resistor R4. Note that TR1 both acts as an *inverter*, and an *interface* between the 5V circuit and the 12V circuit.

### MONOSTABLE

Pins 1 to 6 of IC4 are associated with two NOR gates which act as a monostable. A positive pulse at IC4a pin 1 causes the output pin 3 to go low, and this change of voltage is "conducted" via capacitor C8, making pins 5 and 6 of IC4b also low. Hence the output pin 4 goes high.

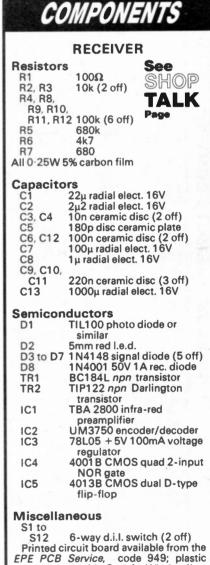
There is now a voltage difference across resistor R5, and so current will flow, charging C8. After a time determined by the

values of R5 and C8, input pins 5 and 6 will be sufficiently positive to be counted as logic 1, and pin 4 switches back to low. The time for which pin 4 is high is not affected by the time for which pin 1 is high, therefore a very reliable pulse is obtained from pin 4 regardless of the pulse time at pin 1.

The output from IC4b pin 4 is fed, via resistor R6, to a Darlington pair TR2 (i.e. two transistors in one transistor package). When these switch on they illuminate l.e.d. D2, and could be used to turn on a relay

etc. if required.

This may be all that is required, depending upon the application for which the remote control circuit is put. However, the circuit diagram Fig. 5. shows two T-type bistables (IC5) and a pair of logic gates (IC4c, IC4d) to provide the more complex control needed for the Curtain Winder described last month.



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### INFRA-RED REMOTE CONTROL - RECEIVER

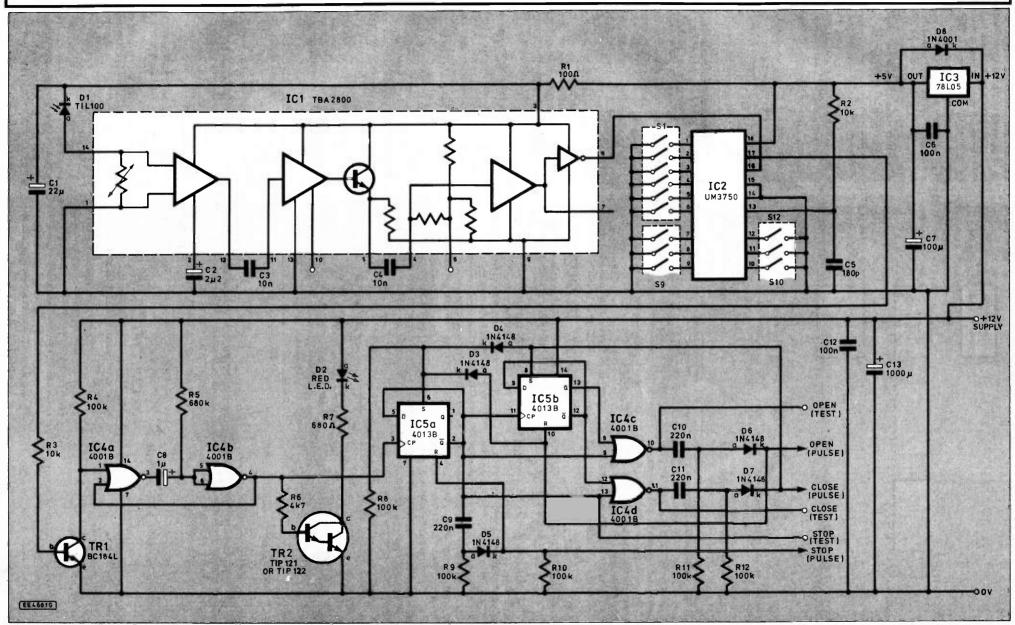


Fig. 5. Complete circuit diagram for the Infra-Red Receiver

### LOGIC CONTROL SYSTEM

The Remote Control System in this circuit offers only one type of signal and hence a single output from IC2. It is possible to design a circuit like a TV Controller with a variety of functions but the single function system offers maximum security since the valid code is one of 4096. It is the job of the logic control system to translate this single command into three outputs suitable for driving a motor in the following sequence

FORWARD/STOP/BACKWARD/STOP If we substitute Open for Forward and Close for Backward the remote control system can be used to control the curtain winder described last month.

### T-BISTABLE

The type of sequence required can be obtained using a type of bistable known as a "T-bistable". It works rather like a toggle action push switch – the type used on many TV sets.

Press it once and the TV switches on, press the switch again and the TV switches off. The T-bistable is an electronic version of this. In simple terms, the first electrical signal switches its output from 0V to positive, the second switches the output back to 0V.

How a T-bistable can be made with the aid of a CMOS 4013 i.c. is shown in Fig. 6. This i.c. actually contains two D-bistables. In brief, a D-bistable has a "data input" (D). It's normal output (known as Q) copies its data input at the instant its "clock input" (CP) is switched from 0V to positive.

A second output (known as "not Q and written Q") assumes the opposite state of Q. The bistable also contains "set" (S) and "reset" (R) inputs. Making Set positive for a moment causes Q to switch to positive; making Reset positive for a moment switches Q back to 0V. This is regardless of the other input pins.

How the  $\overline{Q}$  output can be connected to the "Data" input is shown in Fig. 6. This turns the D-bistable into a T-bistable. We will now describe more fully how the T-bistable works, and will begin by assuming that Q is "low" (0V), and therefore  $\overline{Q}$  is "high" (positive).

Imagine that we switch the clock input from low to high. Since the data input is high (remember  $\overline{Q}$  is high), this will be copied to Q making Q high and  $\overline{Q}$  low. Even though the clock may remain high the outputs will be stable in this new state since data is only copied to Q at the moment when the clock input *changes* from low to high.

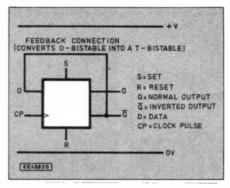


Fig. 6. Using a 4013 i.c. to produce a T-bistable.

Output  $\overline{Q}$  will now be low, making data low. If the clock input is made low, nothing happens. However, when the clock becomes high again, the data input is copied to Q making Q low, and therefore  $\overline{Q}$  high.

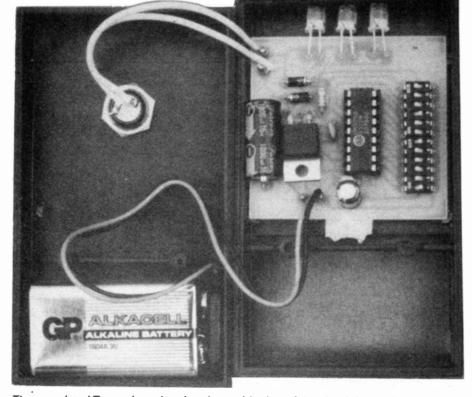
To sum up: First clock pulse causes Q to

switch to high

Second clock pulse causes Q to switch to low

Third clock pulse causes Q to switch to high etc.

Notice that Q switches between low and high at half the rate of the clock pulse. For



The completed Transmitter showing the positioning of the circuit board, battery and "trigger" switch.

this reason the circuit is sometimes called a frequency divider.

### **Practical Note**

When using a D-bistable such as the CMOS 4013B, note that the positive supply is connected to pin 14, and the 0V supply to pin 7. All inputs must be connected to something i.e. not left "open circuit".

### EMPLOYING THE T-BISTABLES

Returning to Fig. 5, IC5 is the CMOS 4013B i.e. This contains two D-bistables. We have made them behave as T-bistables by connecting  $\overline{Q}$  (pin 2 and pin 12) to Data input (i.e. pin 5 and pin 9). In the fully reset state pin 2 will be high, pin 13 low and pin 12 high.

The output from the monostable (pin 4 of IC4) is fed to clock input, pin 3 of IC5. Each time pin 3 (IC5) goes high, pin 1 (Q) changes state. Pin 2 also changes state, but is always at the opposite logic level to pin 1.

When the circuit is in its Reset state, pin 2 is positive. Hence pin 2 is used as the Stop command, and a glance at Fig. 5 reveals pin 2 is directly connected to the Stop Test output. Pins 1 and 2 of IC5 could therefore be used for on/off control since they both change stage at each pulse from the monostable.

However, the curtain winder and similar circuits require a more complex command sequence i.e. Stop/Forward/Stop/Reverse etc. A second T-bistable, together with two logic gates achieve this.

The monostable required two NOR gates, leaving two spare gates in IC4. These two spare gates (IC4c, IC4d) are used to decode the outputs from pins 2, 12 and 13 of IC5.

The output from a NOR gate is "high" only when both inputs are "low". Hence, whenever Stop (pin 2 of IC5) is high, pins 9 and 13 of IC4c/d will be high and output pins 10 and 11 will be low. The other input pins 8 and 12 of IC4c/d are connected to output pins 13 and 12 of IC5. This ensures that when Stop is low only one or other NOR output is high, achieving the sequence, Stop/pin 10 of IC4 high/STOP/Pin 11 of IC4 high etc.

If pin 10 (IC4c) is high, Open is active. If pin 11 (IC4d) is high, Close is active. The outputs from pins 10 and 11 of IC4 are available on the p.c.b. and are labelled Open Test and Close Test. They are useful for testing the circuit since they assume a steady logic level which can be detected using a voltmeter. The Curtain Winder requires pulses rather than steady logic states; hence the need for the capacitors (C10, C11), diodes (D6, D7) and resistors (R11, R12).

### INTERFACING TO THE CURTAIN WINDER

When adding the remote control Receiver to the Automatic Curtain Winder it is important not to interfere with the existing control of the Curtain Winder. It is also essential for the remote logic to respond to the commands issued by the separate manual switches in the winder logic, and the light sensing circuit.

The Open, Close and Stop commands were therefore reduced to short pulses which are fed directly to the appropriate switch inputs on the winder control p.c.b. Diodes D5, D6 and D7 are necessary to block any flow of current from the Curtain Winder logic circuit.

# \* LONGER LEAD \*\* LONGER LEAD

Fig. 7. Transmitter printed circuit board component layout and full size copper/foil master.

At first sight diodes D3, D4, and the feedback connections to the Set and Reset pins (6, 4, 8, 10 of IC5) are superfluous, since the feedback links will try to force the bistables into the states they already hold. However, the purpose of these links is to keep the remote control logic synchronised with commands issued at the Curtain Winder logic end.

For example, if darkness triggers the curtains to close, the fact that the curtains *are* closed must be relayed back to the remote control logic so that at the next press of the remote control button the curtains open.

This transfer of information works very well, enabling the curtains to be controlled by daylight, separate push button switches, the automatic current-sense stop circuit on the Curtain Winder p.c.b., and the Remote Control, all independently and in step. If curtain track stop switches are used to sense when the curtains are fully opened or fully closed there is no feedback to the remote control circuit and the system will not be fully synchronised. For this reason the automatic "current-sense" stop circuit included on the Curtain Winder p.c.b. is recommended if the remote control option is employed.

## TRANSMITTER CONSTRUCTION

The printed circuit board (p.c.b.) component layout and underside copper foil master pattern for the Transmitter is given in Fig. 7. This board is available from the EPE PCB Service, code 948.

designed for a "remote control box", including space for a PP3 battery, size 103mm x 62mm x 23mm. Begin construction by checking that the p.c.b. will fit the space allowed. It is much easier to

The p.c.b.

is

trim the edge of the p.c.b. before the components are mounted.

Following the layout shown in Fig. 7, begin work by soldering in the i.c. socket and smallest components, watching out for the polarities of diodes, transistor and electrolytic capacitors.

### D.I.L. SWITCHES

Switches S1 to S12 were two sets of 6-way d.i.l. switches in the prototype. These are used to determine the code transmitted. If no other similar remote control systems are employed in the house the d.i.l. switches may be omitted.

Alternatively, one or more short wire links may be inserted for certain of the

switches S1 to S12 to create a particular fixed code. It is essential that the same code is used in the Receiver.

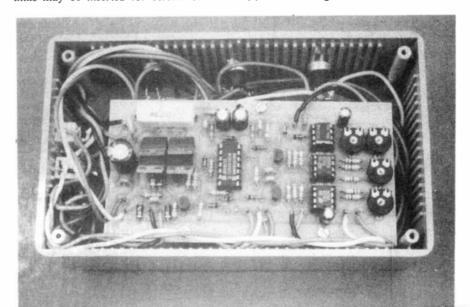
The infra-red l.e.d.s should be soldered in (the correct way round) so that they protrude over the side of the p.c.b.. When the p.c.b. is fixed into the case the l.e.d.s will be in the correct position at the front of the Remote Control Unit.

Fit terminal pins as required, and connect the switch S3 and PP3 battery clip, or wires to the battery terminals. Finally insert the i.c. noting that its notch must be at the top.

Use a 5V or 9V 100mA regulated supply if possible, for testing. Do *NOT* use 12V since the i.c. is only rated to a *maximum* of 11V. If a regulated supply is not available insert a PP3 battery, taking care to fit it the correct way round.

### TRANSMITTER TESTING

It is difficult to fully test the Transmitter until the Receiver is finished. If an early test is essential, an oscilloscope may be used to detect the signal on output pin 17 of IC1. If necessary, trace this signal to the collector (c) of the Darlington transistor TR1.



### RECEIVER CONSTRUCTION

The topside printed circuit board (p.c.b.) component layout and full size underside copper foil master pattern for the Receiver is shown in Fig. 8. This board is also available from the *EPE PCB Service*, code 949.

Remember to check the options available before beginning construction. If you wish to use the circuit with the Automatic Curtain Winder then build the whole circuit.

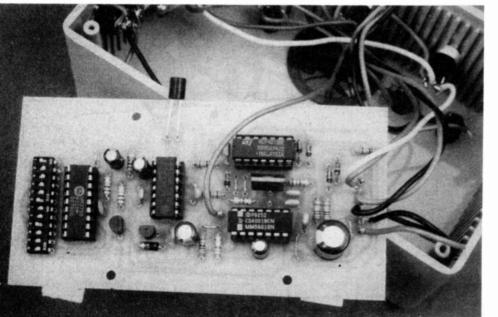
If you only require a relay output, which switches on whenever a pulse is

received, then omit IC5 and associated components, but do not leave the input pins 8, 9, 12, 13 of IC4 open circuit; connect them to the 0V line. Alternatively, if you required a simple toggle action i.e. first pulse turns on, second turns off etc. then use the output from pin 2 of IC5 i.e. the STOP test output.

Begin construction of the Receiver board by soldering in the i.c. sockets, followed by the wire links and smallest components. Take care to fit the diodes the correct way round, and likewise the transistors, voltage regulator IC3 and electrolytic capacitors.

# The close makin helpfus small and to the further to

Fig. 8. Component layout and full size copper foil master for the Receiver p.c.b. The anode lead of D1 is the longer lead. The completed board, wired into last month's Curtain Winder case, is shown below.



### D.I.L. SWITCHES

Switches SI to SI2 are two sets of 6-way d.i.l. switches in the prototype. These are used to determine which code the Receiver is programmed to recognise. If no other similar remote control systems are employed in the house the d.i.l. switches may be omitted.

Alternatively, one or more short wire links may be inserted for certain of the switches SI to SI2 to create a particular fixed code. Clearly it is essential that the same code is used in the transmitter circuit.

### **TERMINAL PINS**

The terminal pins on the p.c.b. are quite close together and the following tip on making connections to these pins may be helpful. First "tin" the pins i.e. place a small amount of solder on each one. Strip and tin the connecting wires so that only 2mm. of bare tinned wire shows.

The connecting wires may now be joined to the terminal pins without the need of further solder. Ignore the old advice about

making a mechanically sound joint which does not rely on the strength of the solder – solder is more than adequate to secure a connecting wire.

Do NOT connect the infra-red

sensor DI at this stage. The red l.e.d. +12V D2 should be connected directly to the appropriate terminal pins the correct way round, or temporarily via wires, for testing. Do not insert the i.c.s at this stage, other than IC3 which should have already been soldered in

### PRELIMINARY TESTING

It is well worth checking the voltages around the circuit before inserting the i.c.s and infra-red sensor D1 since a great deal of expensive damage can be avoided. Connect the circuit to a 12V power supply (a 12V 100mA regulated supply is ideal) and use a voltmeter to check the supply across pin 3 (positive) and pin 9 (negative) of ICl's socket.

The voltmeter should read 5V. If not, check the voltage regulator IC3 carefully. If all is well check that the supply across the IC2 socket is also 5V (pin 18 positive, pin 14 negative).

If all is well connect the infra-red sensor D1. In the prototype circuit the infra-red sensor was soldered directly to the terminal pins, leaving its wires as long as possible so that when flat against the p.c.b. the sensor protrudes over the edge.

Bend it up a little, for testing, and check again that it is connected the correct way round. Now insert the i.c.s checking that their notches are at the top.

### TESTING THE SYSTEM

To check that the system is working, aim the Transmitter at the Receiver (the flat side of the sensor), holding the transmitter at a distance of about one metre. The Receiver should cause l.e.d. D2 to light for a brief time.

If there is no response check that the *code* on the Receiver matches the Transmitter i.e. the same pins are connected to 0V on the UM3750 encoder/decoder i.c.

The Receiver circuit is sensitive to reflections, and reliability can be improved by keeping the sensor away from a reflective surface e.g. by placing a back tube over the front of the sensor. At this stage a lens should not be necessary.

For reliable results the transmitter should be a metre or more from the receiver.

### **FURTHER TESTS**

If the red l.e.d. D2 continues to refuse to work it is important to establish whether the fault lies in the Transmitter or Receiver (or both!). Unfortunately, an oscilloscope is virtually essential at this stage.

Connect the oscilloscope ground to 0V in the *Receiver*. Connect the oscilloscope input to the side of the receiver sensor D1 which joins pin 14 of IC1. When the transmitter is triggered a signal should be apparent on the screen. If no signal is visible, try another remote control transmitter from a TV etc. If a signal now appears on the scope examine your Transmitter carefully.

Assuming that your own Transmitter is working, an amplified signal should be present on pins 12 and 5 of IC1 in the receiver. Finally, a much amplified signal should be present at the output pin 7, and an inverted version at output pin 8.

If no signal is apparent on pin 14 of IC1 (even when a TV or video recorder remote control unit is used), check the Receiver power supply (and IC3) and that the sensor D1 is connected the correct way round. If a signal is detected at pin 14, trace the signal through pins 12, 11, 5, 4 of IC1, and the output signal at pin 8. A dual-beam oscilloscope will allow a direct comparison between the signal at pin 17 of the i.c. in the Transmitter, and at pin 8 of IC1 in the Receiver.

A valid code received at pin 16 of IC2 in the Receiver should make the voltage at pin 17 (which is normally at about 5V) fall to 0V. This fall may be very brief and difficult to measure on a voltmeter.

The monostable should respond making pin 4 of IC4 switch from 0V to about 12V for about half a second. If necessary the monostable can be tested by shorting pin 3 to 0V for a moment. This should make pin 4 (IC4) switch to 12V, and trigger the Darlington TR2, causing, D2 to light for about half a second.

Once it is established that the output of the monostable (pin 4 IC4b) switches to 12V for a short time, check that pin 3 of IC5 copies this. Testing of the logic section of the circuit can be done by connecting a voltmeter (with its negative lead connected to 0V in the circuit) to the Open (test), Close (test) and Stop (test) points in turn.

The Open (pulse), Close (pulse) and STOP (Pulse) outputs designed to be connected to the Curtain Winder circuit can only be tested properly using an oscilloscope, since the voltage will pulse for a very brief time.

### INTERFACING TO OTHER CIRCUITS

Clearly a remote control circuit is designed to connect to other equipment. A relay could be connected from positive to the collector (c) of the Darlington TR2, not forgetting to add a protective diode connected across the relay coil, with the diode's cathode (k) towards the positive supply.

A Darlington pair module could be connected to the STOP (Test) output to provide a "toggle" action i.e. ON/OFF/ON/OFF each time the Remote Control button is pressed.

### **CURTAIN WINDER**

To link the Receiver to the Curtain Winder you should connect the Open (Pulse) output from the remote control receiver to the Open pushbutton switch input on the Curtain Winder p.c.b.. Do the

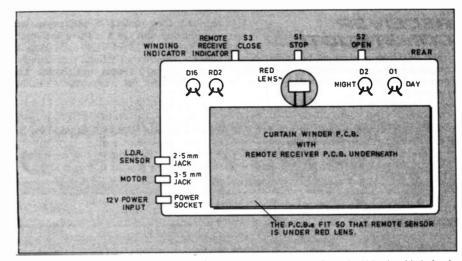


Fig. 9. If the receiver is incorporated into last month's Curtain Winder Unit it sits underneath the Winder p.c.b. so that the i.r. sensor aligns with the lens "window".

same with Close (pulse), and STOP (pulse) to their respective inputs on the Curtain Winder.

Normally both the Remote Control Receiver and Curtain Winder p.c.b.s will share the same 12V supply. If this is not the case it is important to connect 0V on one p.c.b. to 0V on the other.

# HOUSING THE SYSTEM

The Transmitter p.c.b. should fit exactly into the remote control case as shown in the photographs. Begin by drilling holes for the infra-red diodes (D2 to D4), and pushswitch S13. The use of p.c.b. supports may be helpful, but they may make the p.c.b. foul the top of the case. In this instance the p.c.b. will be such a tight fit that additional support is unnecessary.

The Receiver p.c.b. is similar in size to the Curtain Winder p.c.b. and will fit very neatly into the specified case (64.5mm × 129.5mm × 42.5mm external dimensions). Fig. 9 shows how the Curtain Winder p.c.b. is mounted piggy-back fashion on the Remote Receiver p.c.b. Arrange them this way round so that the presets on the curtain winder are accessible.

The use of p.v.c. spacers and self-tapping screws can be used to clamp the boards together, the Receiver p.c.b. being mounted to the case using self-adhesive p.c.b. supports. Alternatively, single M2·5 nuts and bolts could be used with short and long spacers to fasten the pair of p.c.b.s in three or four places. Check exactly where they will be fitted, and then mark and prepare the case for drilling.

Holes will be needed for the three curtain control push switches, four l.e.d.s, a large hole for the infra-red sensor, and holes at the side for a power input socket, 2.5mm. jack socket for the LDR and 3.5mm. jack socket for the motor (see photographs).

Drilling should present no difficulty, but the infra-red sensor hole needs to be about 20mm in diameter assuming that a red plastic lens is used. This lens is flat on one side so that it can be glued with its flat surface against the inside of the case. It may be necessary to trim the top of the lens using a saw or cutters. Use of the lens should increase the operating distance of the system although the difference may be marginal; if in doubt, experiment for the best effect.

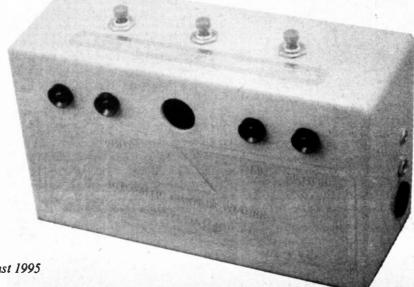
Fix the switches, l.e.d.s and sockets in position, and after gluing in the lens complete all the external connections using a variety of colour coded wires, noting the importance of connecting the l.e.d.s and power socket with the correct polarity.

If possible, connect the motor temporarily, so that its polarity can be reversed if necessary. It may be difficult to know before the first trial which way the motor will turn.

### FINAL TESTING

Hopefully the Curtain Winder section of the project will already be functioning, and the mechanical part of the unit which was described last month. All that remains is to connect the motor, connect the LDR if required, and connect the 12V supply (a 12V 1A regulated mains adaptor is ideal).

The Remote Control Unit should cause its l.e.d. D2 to pulse when the message is received, and this in turn should wind or stop the curtains.



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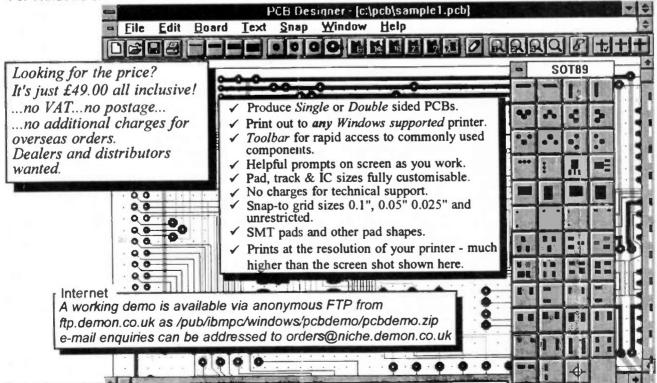
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# INTERFACE

## Robert Penfold

THE decoding of amateur RTTY (radio teletype) and CW (Morse) signals has been covered in previous articles in *EPE*. The systems described previously relied heavily on hardware add-ons to do the decoding, with the computer having to do little more than print the received characters on the screen. In recent years a different approach has become quite popular, and this method relies on the software to provide virtually all the signal processing.

One probable reason for the increased popularity of this method is the availability of a shareware program for PCs called "Hamcomm 3-0". This program was written by W.F. Schroeder (DL5YEC), and although it is German in origin, it is supplied complete with an excellent on-disk English manual. It can be used for thirty days free of charge, after which a \$30 registration fee must be paid if you wish to continue using the program. As \$30 is currently equivalent to about £19, this represents very good value for what is quite a complex piece of specialised software.

### **Minimalist**

While "Hamcomm" does not totally avoid the need for a hardware interface between the audio output of the receiver and the input port of the computer, it permits a minimalist approach. The manual includes the circuit diagram for a suggested interface, which is basically just an open loop operational amplifier which clips the audio output from the receiver and drives an input line of a serial port.

If preferred, the program can be used with a conventional tone decoder, but some of its facilities will then be lost. In particular, it will not provide any help with accurate tuning, and determining the shift of an RTTY signal. The program can handle transmission incidentally, and again needs only a very basic interface. However, I have not explored this aspect of the program, and it is not something we will consider further here.

The conventional approach to RTTY/CW decoding uses filters, level detectors, phase locked loops, etc. to recover the basic digital signal. In the case of RTTY, a UART (universal asynchronous receiver/transmitter) then provides the serial to parallel conversion. All the computer then has to do is convert the five bit Baudot codes to ASCII characters that are then displayed on the screen. Converting the dots and dashes of Morse code into ASCII characters is a bit more difficult, even if the hardware sorts out the dots from the dashes. It is still a relatively simple process though, with the hardware doing most of the processing.

A program such as "Hamcomm" avoids the need for a lot of expensive hardware, but it has to provide all the processing, including the tone decoding. An RTTY signal uses a conventional f.s.k. (frequency shift keying) system, where the two logic levels are represented by two audio tones. A tone spacing of 170Hz is normally used for amateur transmissions, but commercial stations mostly use a wider spacing of 425Hz or 850Hz. In order to determine the frequency of the input signal the computer measures the time between transitions on the input line. In this way the program can tell which frequency of an RTTY signal is being received, and whether or not the correct tone for a CW signal is present.

### Rapid Demand

This type of program is fairly demanding on the computer, since it takes some rapid processing in order to accurately analyse the input signal and determine the frequencies present. Apart from the fact that this has to be undertaken in real-time, the computer must also carry out software routines to provide serial-to-parallel conversion, and to print the decoded characters on the screen. Again, this must be done in real-time.

Although "Hamcomm" uses efficient programming, it is necessary to use some form of AT PC in order to guarantee it will run fast enough. Its more advanced features benefit from a faster computer, but I found that it ran reasonably well on a 33MHz 80386 based PC. In other respects it is not very demanding. The minimum requirements are 374K of free memory, MS-DOS 3·0 or higher, and any standard graphics adaptor if you wish to use the more advanced features of the program. The

program will detect any standard video

adaptor, and configure itself accordingly.

### Interface

The manual provides the circuit for a basic interface, but I used my own design (Fig.1). A basic clipping amplifier will presumably work well enough if the short wave receiver has built-in filtering for RTTY and CW reception. With most receivers it is necessary to make do with the SSB (single sideband) mode, which will give a much wider bandwidth than is needed for CW and most RTTY reception. This often results in the computer being presented with a plethora of signals, making it difficult for the software to sort out the wanted signal from the general hubbub of signals.

The clipping amplifier is IC3, and it is preceded by two fourth order (24dB per octave) filters. The filter based on IC2a provides lowpass filtering with a cutoff frequency at a little over 1kHz. IC2b is used in a highpass filter which has a cutoff frequency at about 500Hz. This gives a reasonably flat passband about 500Hz wide, which is sufficient to accommodate CW signals, plus RTTY types having a tone shift of 170Hz or 425Hz. Switch S1 should be closed when receiving RTTY signals that have a tone shift of 850Hz. Switch S1 then bypasses the filter components and removes the filtering. IC1 simply acts as a buffer stage at the input of the circuit.

By the standards of short wave receivers the extra selectivity of the filter is not particularly great. Nevertheless, it seems to

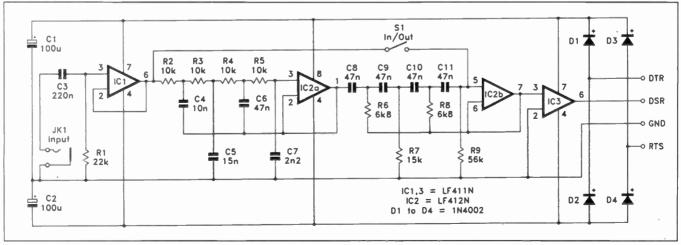


Fig. 1. Circuit diagram for the CW/RTTY interface.

make a worthwhile improvement in performance. The "Hamcomm" manual warns against using highly selective audio filtering. Apparently this can "tune" the background noise, making it difficult for the program to determine whether or not a tone is present.

Supply Problems

The circuit is powered from the otherwise unused DTR and RTS outputs of the serial port. Diodes D1 to D4 provide bridge rectification so that the interface receives a supply of the correct polarity regardless of the logic level on DTR and RTS lines. C1 and C2 are smoothing/decoupling capacitors. Due to the positive/negative signal voltages of an RS232C interface, the circuit is provided with dual balanced supplies. Only a very limited supply current is available, and low current operational amplifiers have therefore been used throughout the circuit. The total current consumption is slightly under one milliamp, plus any output current drawn from IC3. The serial port should be able to supply this without difficulty. I would not recommend the use of ordinary operational amplifiers such as the LF351N and LF353N.

The serial port can be a standard 25-pin type, or a 9-pin AT type. The connection details are shown in Fig. 2. In either case a female D-type connector is needed to make the connections to the computer. The input of the interface is fed from an audio output of the receiver via a screened cable fitted with the appropriate audio connectors. Where possible it is better to use the "tape" output of the receiver. This gives an output level which does not change when the volume control is adjusted, and it leaves the loudspeaker and headphone socket available for normal use.

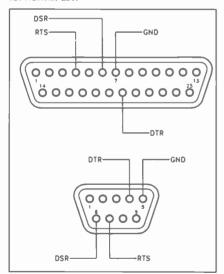
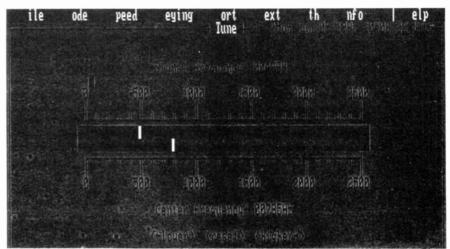


Fig. 2. Connections for 25-pin and 9-pin PC serial ports.

It is possible to use the output from a headphone socket, but this will probably cut off the receiver's built-in loudspeaker. It is then necessary to improvise some means of hearing the receiver's audio signal. A splitter cable represents an easy way of driving both the interface and a pair of headphones. Another possibility is to connect a crystal earphone to the interface between IC2 pin 7 and the earth rail. This enables the filtered audio signal to be monitored. There



Part of the "Hamcomm 3.0" tuning screen.

is insufficient drive available for any other type of earphone or headphones.

"Hamcomm" is very straightforward to install and use. The standard screen arrangement has a menu bar at the top, with the main screen split in two. The bottom section of the screen is used to print received data, and (where appropriate) the top section of the screen is used to show data that has been typed in for transmission. The program can be controlled via a mouse and popdown menus, or via the keyboard.

I only used "Hamcomm" to decode amateur RTTY and CW signals, but it can also handle seven and eight bit ASCII, AMTOR, and SITOR A/B. The usual baud rates and f.s.k. shifts can be accommodated, and non-standard shifts can be set if necessary. Pressing the TAB key toggles the program between normal and reversed polarity, so it is still possible to decode an RTTY signal if the receiver is set to the wrong sideband mode.

For any system of this type to work well it is essential for the tuning to be accurate. Pressing function key nine switches the program to the tuning mode. This brings up a horizontal scale across the middle of the screen calibrated in Hertz. A marker shows the centre frequency of the decoder, but this can be moved to any desired frequency using the left and right cursor keys. Further markers indicate the frequencies of received input signals.

Tuning to a CW signal is just a matter of adjusting the receiver to line up its on-screen marker with the centre frequency marker. For an RTTY signal the two onscreen markers must be equally spaced either side of the centre frequency marker. Where necessary, the frequency shift of the signal can be measured quite accurately against the frequency scale. There are further facilities available which enable the input signal to be analyzed and displayed in various ways, but this basic tuning facility is all that is needed in normal use.

### Results

A good filter or phase locked loop decoder can correctly decode RTTY signals that are barely audible amongst the back-ground noise. This system seems to be rather less able at decoding signals under difficult conditions, but it is remarkably good when the minimal amount of interface hardware is taken into account. A number of stations around 14.090MHz on the 20 metre band were received quite well, although there were the inevitable problems with noise and fading from time to time.

Results when used to decode amateur CW signals were less good. This was probably due in part to crowded band conditions, with the decoder often being fed with several signals on slightly different frequencies. There is another problem though, which is simply that human generated Morse code often has rather inaccurate and inconsistent timing, making it difficult for a computer to decode it. Some signals that were heavily infested with QRM were decoded quite accurately, while others that were largely free from interference were decoded with very poor accuracy. The rather cryptic nature of amateur Morse code signals does not help matters. The odd character wrong here and there can make complete nonsense of a message. However, it was usually possible to reliably decode the call-signs, plus some other information such as the location (QTH) of the station.

### **Big Noise**

Unfortunately, computer systems are very good at generating r.f.i. (radio frequency interference) at frequencies right across the short wave bands. The problem seems to be more due to r.f.i. from the monitor than the computer itself. A modern low radiation monitor is likely to be less troublesome than an older type which is largely void of any electrical screening.

Using a short indoor aerial is unlikely to provide usable results. A long outdoor aerial will provide about the same amount of noise but other signals should be much stronger. The best type of aerial is an outdoor type mounted well away from the computer, with a screened or balanced feeder to connect it to the receiver. This should give strong signals with minimal pickup of noise from the

computer system.

Any reasonably sophisticated communications receiver should give good results. Modern types having synthesised tuning are preferable to older types though, since good frequency stability is a decided asset for CW and RTTY reception. "Hamcomm 3.0" is available from The PDSL, Winscombe House, Beacon Road, Crowborough, Sussex, TN6 1UL (Tel. 01892 663298, Fax 01892 667 473).

Mixed metal/carbon film resistors ** W E12 series 10 ohms to 1 Megohm	2р
Carbon Film resistors ¼W 5% E24 series 0.51 R to 10MO	1р
100 off per value – 75p. even hundreds per value totalling 1000£6	q00.6
Metal Film resistors 1/W 10R to 1 MO 5% E12 series - 11/p. 1% E24 series	2р
Mixed metal/carbon film resistors ½W E24 series 1RO to 10MO	1½p
1 watt mixed metal/Carbon Film 5% E12 series 4R7 to 10 Megohms	5p
Linear Carbon pre-sets 100mW and ¼W 100R to 2M2 E6 series	7р
Miniature polyster capacitors 250V working for vertical mounting	
.015, .022, .033, .047, .068-4p. 0.1 - 5p. 0.12, 0.15, 0.22 - 6p. 0.47 - 8p. 0.68 - 8p. 1.0 - 12	2p
Mylar (polyester) capacitors 100V working E12 series vertical mounting 1000p to 8200p - 3p01 to .068 - 4p. 0.1 - 5p. 0.12, 0.15, 0.22 - 6p. 0.47/50V - 8p	
1000p to 8200p - 3p01 to .068 - 4p. 0.1 - 5p. 0.12, 0.15, 0.22 - 6p. 0.47/50V - 8p	
Submin ceramic plate capacitors 100V wkg vertical mountings. E12 series	
2% 1.8pf to 47pf - 3p. 2% 56pf to 330pf - 4p. 10% 390p - 4700p Disc/plate ceramics 50V E12 series 1PO to 1000P, E6 Series 1500P to 47000P	4p
Disc/plate ceramics 50V E12 series 1PO to 1000P, E6 Series 1500P to 47000P	2р
Polystyrene capacitors 63V working E12 series long axial wires	
10pf to 820pf - 5p. 1000pf to 10,000pf - 6p. 12,000pf	7p
741 Op Amp - 20p. 555 Timer - 20p. LM3900	. 80p
CMOS 4001 - 20p. 4011 - 22p. 4017 - 40p. 4069UB unbuffered	. 20p
DIL holders, 8-pin 9p; 14-, 16-, 18-pin 12p; 24-pln 18p; 28-pin 20p; 40-pin 25p.	
ALUMINIUM ELECTROLYTICS (Mfds/Volts)	_
1/50, 2.2/50, 4.7/50, 10/25, 10/50	5p
22/16, 22/25, 22/50, 33/16, 4//16, 4//25, 4//50	бр
1/50, 2.2/50, 4.7/50, 10/25, 10/50 22/16, 22/25, 22/50, 33/16, 47/16, 47/25, 47/50 100/16, 100/25 7p, 100/50 220/16 8p, 220/25, 220/50 10p, 470/16, 470/25	. 12p
220/16 8p; 220/25, 220/50 10p; 4/0/16, 4/0/25	. 11p
1000/25 25D; 1000/35, 2200/25 35D; 4700/25	. /0p
Subminiature, tantalum bead electrolytics (Mfds/Volts)	
0.1, 0.22, 0.47, 1.0, 2.2, 3.3 @ 35V – 4.7/16, 6.8/10, 10/6, 10p; 6.8/35, 12p.	
4.7/25, 6.8/16, 10/6, 11p; 15/16, 22/6, 33/10, 15p; 10/25, 16p; 10/35, 22/16, 20p. 47/10, 20p; 47/16, 25p; 47/20, 30p; 47/35, 32p; 100/3, 18p; 100/6, 220/6, 20p.	
4//10, 20p; 4//16, 25p; 4//20, 30p; 4//35, 32p; 100/3, 18p; 100/6, 220/6, 20p.	
VOLTAGE REGULATORS	
1A + or - 5V, 8V, 12V, 15V, 18V & 24V - 55p. 100mA, 5.8, 12, 15, V +	. 30p
DIODES (piv/amps)	0 -
75/25mA 1N4148 2p. 800/1A 1N4006 4½p. 400/3A 1N5404 14p. 115/15mA 0A91	8p
100/1A 1N4002 3½p, 1000/1A 1N4007 5p, 60/1.5A S1M1 5p, 100/1A bridge 400/1A 1N4004 4p, 1250/1A BY 127 10p, 30/150mA 0A47 gold bonded	. 25p
70071A 114004 4p. 1250/1A 61 127 10p. 30/150/1A GOID conded	. 16p
Zener diodes E24 series 3V3 to 33V 400mW - 6p. 1 watt	. 1 Up
Battery snaps for PP3 - 7p for PP9 L.E.D. s 3mm, & 5mm, Red, Green, Yellow - 10p, Grommets 3mm - 2p, 5mm	. 12p
L.E.D. Sonini. a onini. Ned, dreen, reliow - Top, grommets omm - 2p. omm	Zp
Red flashing L.E.D.'s require 9-12V supply only, 5mm	. SUP
Mains indicator neons with 220k resistor	. IUp
High enough of the 1.2 1.2 2.00 pt. Assure 10. Monters, chassis, mounting	op
High speed pc drill 0.8, 1.0, 1.3, 1.5, 2.0mm - 40p. Machines 12V dc	5.00
AA/MPT Nigord sopharecoble cells 2000 control University Market 1005	4.50
AA/HP7 Nicad rechargeable cells 90p each. Universal charger unit	0.50
Class and exist about the control packs of 4	pack
Glass reed switches with single pole make contacts - 8p. Magnets	. ZUP
Jack plugs 2.5 & 3.5m – 14p; Sockets Panel Mtg. 2.5 & 3.5m	. /UP
Jock progs 2.5 d. 3.5m – 14p, Sockets Faher Mtg. 2.5 d. 3.5m	up
Ear pieces 2.5 & 3.5mm, dynamic – 20p; 3.5mm crystal	1.50
Multi cored solder, 22G – 8p yard, 18G – 14p yard.	
TRANSISTORS	
8C107/8/9 - 12p. BC547/8/9 - 8p. BC557/8/9 - 8p. BC182, 182L, BC183, 183L.	
BC184, 184L, BC212, 212L - 10p.	
BC327, 337, 337L - 12p. BC727, 737 - 12p. BD135/6/7/8/9 - 25p. BCY70 - 18p.	
BFY50/51/52 - 20p.	
8FX88 - 15p, 2N3055 - 55p, TIP31, 32 - 30p, TIP41, 42 - 40p, 8U208A - £1.50, BF195, 197 - 1	2p
Ionisers with seven year guarantee, 240V AC, list price £16.95 or more£1	2.50
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### BART TREPAK

"Dial-a-time" with this Up/Down timer. From seconds up to 9 hours 59 mins. With relay switching output it will also control mains appliances. Ideal for darkroom process timing, games timing and elapsed time indicator.

HEN it comes to generating a time delay of a few seconds or even minutes it is hard to beat the trusty old 555 timer for simplicity and accuracy. Even when longer time delays are required such as a few hours, the 555 followed by a binary divider or one of the similar chips such as the ZN1034 offer the same simplicity and get around the problem of using large capacitor values.

The situation is far more complicated however if a variable timer is required. There are not many uses for a fixed 60 second or 60 minute timer but a variable one with a range of one to 60 seconds or one to 60 minutes or even hours is far more

useful.

Timing anything from a process in a darkroom to charging a battery are things which instantly spring to mind while a little more thought reveals more uses such as a games timer or to remind you to do something such as ring someone back in ten minutes.

### **GET SET**

While in theory it is possible to make a timer for these applications using a 555 timer i.e., by making the timing resistor a variable and switching capacitors for different ranges, in practice the problem of calibrating the potentiometer would be difficult to say the least and with the resulting scale, setting a delay with any reasonable accuracy virtually impossible. With some of the devices which use an oscillator and a divider, it is possible to set the time delay by means of switches which are set in binary to give the required division ratio.

The accuracy is then dependent on the accuracy of the oscillator but setting the delay is still not that easy – what is 53 in binary for example. Anyone other than a computer programmer would probably need a bit of time to work it out while most people without a technical background would probably not know where to begin.

For this reason, commercial electronic timers are digital for accuracy and are made to be programmed in decimal to make them easy to set. These employ a

digital display on which the required delay is set and the timer then counts down to zero operating a buzzer when this is reached.

This, of course, is much more complex in actual circuitry than a 555 timer but is much easier to use. Most timers of this type are intended for use in the kitchen or similar situations so they use a small display and do no more than sound a buzzer to signal the completion of the timing period.

### **PLAY TIME**

This project began life because of a need for a timer to put an end to the not so friendly arguments about when indoor games should have finished. The original requirement was for a timer which could be easily set to time for up to 20 minutes using the display, i.e. no binary, or BCD coded thumbwheel switches.

It also had to have a Start/Stop button so that the timing could be suspended for any reason and then restarted, and a buzzer which would sound at the end of the game. It was also felt more desirable (for some reason) to display the time elapsed rather than the time remaining so the unit would have to count up rather than down.

While working on the design, it was soon realised that no dedicated i.c.s were available to perform this function and a microcontroller would have to be used, so various other features were considered to make the finished unit more versatile and useful in other applications. The specification was therefore "expanded" to a 3-digit timer which could count in minutes and seconds as well as in hours and minutes, count up or down and provide an output for switching off a mains appliance at the end of the period.

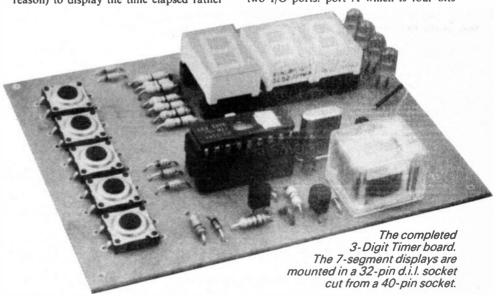
The additional cost in providing these features is negligible, requiring as it does only a few extra lines of program code and a relay. It also resulted in a timer with many more applications than the one with

the original specification.

### FOUR INTO TWO WILL GO

The first step in any microcontroller design, after deciding what the device is to do and what functions are going to be required (i.e. the number of switches, displays etc.) is to assign the available input/output lines to the peripheral devices.

The PIC16C54 microcontroller chosen for this design (because of its low cost) has two I/O ports: port A which is four bits



wide and port B which is eight bits, giving a total of 12 lines which can be made to function as inputs or outputs under program control. The pinout line-up and function is indicated in Fig. 1. Two of the lines of port A were chosen to drive the output relay and the buzzer (A0 and A1 respectively) and seven lines (port B lines B0 to B6) were required to drive the 7-segment l.e.d. display (see Fig. 6.).

Since only five function switches are required, five of the seven segment lines could double as drivers for the switches taking each switch contact high in turn and sensing the logic level on the input (port B line B7) which will be high if the switch is pressed. This line has to be dedicated to this function because if one of the segment lines were used for the input line, pressing a switch would short two of the segment lines together causing spurious segments to light or even damage the microcontroller i.c.

### DRIVER

As with all multiplexed displays, the segment drives for all the l.e.d. Digit Displays are common and each display must be driven in turn by a separate digit drive line when the correct segment lines for that digit have been energised. In this design there are effectively four digits; the three 7-segment l.e.d. displays and the indicator l.e.d.s which show which mode (hours/mins or mins/secs and up or down counting) has been selected.

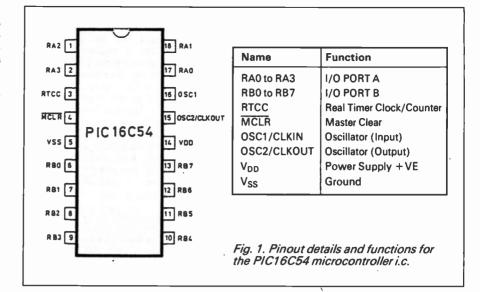
This would normally require four lines but we have only two (A2 and A3) left out of the twelve. Fortunately, the PIC I/O lines enable us to drive two digits with one line.

This is accomplished by using a com-mon Anode and common Cathode display connected in inverse parallel and driving the digit line high and segment lines low to energise the common anode l.e.d. and digit line low and segment lines high to light the common cathode unit. When both units have to be off when the other digits are being driven or the switches are being interrogated, the digit line is put into its high impedance (input) mode, preventing either unit from lighting. Although you will not require sun-glasses to view the resulting display, assuming reasonably efficient l.e.d.s are used, the brightness should be sufficient for most indoor applications. The use of a red filter will also enhance the brightness.

The problem with driving l.e.d. 7-segment displays in this way, is the low reverse voltage rating of the diodes used in the display. This can lead to a segment passing sufficient current when it is reverse biased to light a segment in an adjacent digit.

No damage can result from this because the current is so low and if the segment in the next digit is meant to be on anyway, there will be no noticeable problem. If the segment is meant to be off however, it may be energised at a low level giving a "ghosting" effect.

One way of solving this problem would be to insert a silicon diode, which has a much higher reverse voltage rating, in series with each l.e.d. but this would result in having to fit 25 extra diodes to the circuit which would not be a very elegant solution. Since the reverse voltage rating of l.e.d.s is about 3.5V, driving them from a supply voltage of less than this would circumvent the problem. This is also the minimum microcontroller supply voltage and so to give a larger margin, a diode has been inserted into each digit drive line (four diodes) and the supply voltage increased to 4V.



### TIMEBASE

The most important thing about a timer is that it keeps time and so the next thing to consider is how this is to be done. The PIC (in common with many other microprocessors) contains a counter/timer (RTCC or Real Time Clock/Counter as it is called in the data sheet) which can be incremented by the internal instruction clock which is simply the crystal oscillator frequency divided by four by an internal divider.

The RTCC is an 8-bit counter which can count to a maximum of FF hex (255 decimal) and when it receives the next pulse, will "roll over" to 00 (zero) and start again. By selecting a suitable crystal frequency and loading the RTCC with a suitable value, the zero state would be reached every say 20ms which could then be further counted to provide us with seconds, minutes etc.

Many micros are designed to generate an interrupt signal automatically when the internal counter/timer reaches the zero state allowing the processor to continue executing the main program until this occurs. Unfortunately, the PIC16C54 is not one of them. The only way to see if the RTCC register has reached zero is to read it each time it is incremented.

This would, of course, mean that the microcontroller would have a full time job just reading the RTCC register because it is incremented at the same rate as the instructions are executed which would leave no time to do anything else. For this reason, the PIC has a prescaler which can be used to divide the instruction clock by up to 256 if required allowing the RTCC register to be incremented at an even slower rate.

If we divide the instruction clock by say 100 and load the RTCC with a value of 246 decimal so that it will reach 256 (or zero) after 10 input pulses, then the PIC will be able to execute  $10 \times 100 = 1000$  instructions between each zero state of the RTCC. Thus following an RTCC zero, we can go off and run a program up to 1000 instructions long before returning to check the RTCC again, confident in the knowledge that the zero state will not have been reached by the time we return.

Ideally we want to choose a crystal which will give a nice round figure for the timebase when it is divided by the prescaler and RTCC such as 10ms, 20ms or 50ms and not 19.531ms or some other odd number. While this could be used, we would have to keep adjusting the count

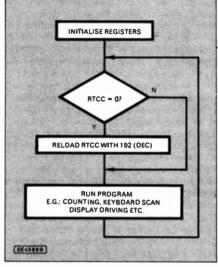


Fig. 2. Basic flowchart.

every second or minute otherwise our timer could be out by quite a few minutes after a long timing run.

The other consideration is that in between the times that the RTCC counter reaches zero, among other things the l.e.d. display must be updated. This must be done sufficiently often to make the display appear constant and not show an annoying flicker. As well as this, there must also be sufficient time between the RTCC "zeros" to enable all the other instructions to be executed.

To obtain a flicker free display, the l.e.d.s will have to be updated at least every 20ms and a 3·2768MHz crystal can be divided to give this figure quite easily. The PIC divides the crystal frequency by four to obtain the instruction clock (and also the frequency at which the prescaler will count) which gives us a frequency of 8192kHz or a time of 1·2µs per instruction.

In 20ms therefore, we can execute over 16,000 instructions which is more than enough to execute the whole program many times over. In fact the final program "wastes" most of these carrying out a delay loop so that the pulse width of the l.e.d. drives is longer to obtain a brighter display!

If we set the prescaler to divide by 256, the RTCC will be incremented at a frequency of 3200Hz and by loading the RTCC with 192 decimal, zero will be reached after 256-192=64 increments, giving a period of 64/3200=0.02s or 20ms.

### DESIGNING THE FLOWCHART

We now know roughly how the program will work so we could draw the flowchart shown in Fig 2. This shows the Timing section of the program with the rest of the routine called Program for the time being. This will finally include the display driving routines, keyboard reading etc.

Before this program can be executed however, we must first set up the microcontroller so that when the circuit is first switched on, it will be in a certain condition with the displays set to zero for example, not counting and in the count down mode. These conditions along with such routine things as defining which ports will be inputs and which outputs, the division ratio of the prescaler etc. can be set up in that part of the program marked Initialise which will be executed once when the circuit is first powered up.

Designing a flowchart is simply a process of drawing more and more detailed charts showing the logical flow of the program until it is possible to write down the instructions as defined in the instruction set of the microcomputer being used. The more familiar the programmer is with the instruction set and the more experienced he is in programming, the earlier this step can be performed and the fewer the number of flowcharts which need to be drawn.

Many programmers can, especially in more simple programs, dispense with the flowchart altogether. This is not to be recommended because, if nothing else, even a simple flowchart can give an overview of the way that a program works which can save many hours work in the future should a program need to be modified.

In Fig. 2 the RTCC counter will overflow every 20ms but the l.e.d. display will need to be updated every second or every minute depending on the mode selected and the l.e.d. indicating that the unit is counting (which flashes every second) will need to be switched on and off every half second. The RTCC "overflows" will therefore need to be counted to determine when the l.e.d. needs to be turned on and off and the resulting half-seconds counted to determine when a second or minute has elapsed.

### **EXPANSION**

An expanded flowchart which now includes a Half-Second counter, which counts to 25 (decimal), is shown in Fig. 3. If the circuit is not in the Run mode (i.e. not timing) then the program branches at the third decision box, switches off the relay and goes on to scan the keyboard and display before returning to test the RTCC again until the next overflow is detected, but causing the l.e.d. to remain on permanently.

If the circuit is in the Run mode, then the relay is switched on and the l.e.d. is complemented (i.e. switched on if it is off and off if it is on) and another counter, (the Timebase counter) is decremented each time the Half-Second counter reaches zero.

The Timebase counter is loaded with the number contained in a Timebase register which is loaded with a value of 2 or 120 (decimal) depending on whether the hours/minutes or minutes/seconds mode has been selected. The timebase counter will therefore reach zero after two half-seconds (i.e. each second) or 120 half-seconds (i.e. 60 seconds or each minute) and this is used to increment or decrement the display.

Note that the read/write (or RAM)

memory in the PIC microcontroller is organised as a set of 8-bit registers which can be operated on by the various instructions. Thus a register may have individual bits altered from 0 to 1 or may be incremented or decremented like a counter. In this article, the term "counter" is used to denote a register used for counting while a "register" simply holds a numerical value or remembers the state of an output but both are in fact identical.

Although the timing chain looks complicated, it does in fact contain very few instructions (each box being typically one or at most two instructions in PIC assembler language) and so it is ex-

ecuted very quickly. Even the boxes marked "Increment/Decrement Displays" and "Keyboard Scan" contain relatively few instructions and execute quickly.

Rather than have the processor running the RTCC test routine all the time with short breaks to execute the rest of the program, a delay routine has been incorporated into the subroutine. Marked "display", this ensures that the digits are driven for as long as possible during the 20ms interval between RTCC zeros to give a brighter display while still leaving enough time to execute the rest of the program.

A delay routine simply sets up a dummy

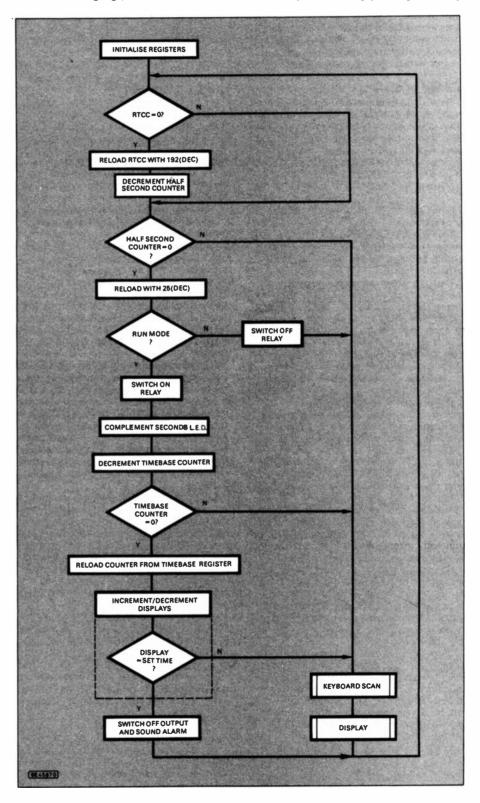


Fig. 3. Expanded flowchart for the timer, including a half-second counter.

counter which does nothing but counts down to zero and uses up time. The larger the number which is loaded into the counter at the beginning of the routine, the longer the time taken to reach zero and therefore the longer the delay.

### JUST ROUTINE

The next step in the design is to expand further the boxes marked Initialise, Increment/Decrement Displays, Keyboard Scan and Display. Rather than describing all of these in detail only the first two will be dealt with and only the functions of the other subroutines mentioned.

The Initialise routine is shown in Fig. 4 and, as mentioned earlier, is only executed once shortly after the unit is first powered up. It consists mainly of instructions loading values into the various registers and counters. One trick which is used to save on switches and make for a less complicated front panel is to read the status of a switch during this time and if it is pressed, to use this to change the function of the circuit.

Since this timer is to count up or down, this could be set up at switch-on as it is unlikely that the user will want to change this during a timing operation. The program therefore reads the Start/Stop switch and if this is pressed when the unit is switched on, the circuit is placed into the

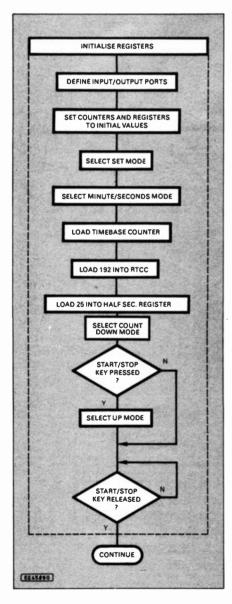


Fig. 4. Initialise routine flowchart.

UP mode and an l.e.d. lit to indicate this.

For DOWN mode the circuit should be switched on without depressing the switch. So that the main program is not entered and the switch read as Start/Stop as it would be in the Keyboard Scan routine, the program is made to wait until the key is released before continuing.

The Increment/Decrement Display routine is shown in Fig. 5. This routine has two outputs or exits: Continue Timing – which goes on to Keyboard Scan and Display, and Timing Completed – which ends the timing cycle and goes on to switch off the relay, sound the alarm and set up the unit for another timing run. Depending on whether the UP or DOWN mode has been selected, either the right hand or left hand side of the flowchart will be followed.

The time is counted on two counters called Tens/Unifs and Hundreds. These are simply memory locations or registers in the PIC which consist of eight bits with the Tens being the top four bits and Units the bottom four bits of one register and the Hundreds the bottom four of another. The Tens/Units are therefore stored as two BCD (binary coded decimal) numbers in one register and the Hundreds stored as a single BCD number in another.

In the DOWN mode, the counters are loaded with the values stored in a similar set of registers (which are incremented by means of the Set Units, Set Tens and Set Hours/Minutes switches) when the Run switch is pressed. Each time this routine is executed, the Tens/Units counter is decremented and when it reaches zero, it "overflows" to FF (hex) (255 decimal). The Hundreds counter is then checked and if this is already zero then the timing is over but if it is not, the Hundreds counter is decremented and the Tens/Units counter loaded with 59 to continue counting down from there.

If the Tens/Units counter does not equal FF(hex) after it has been decremented, then the lower four bits are checked to see if they have not "overflowed" past zero giving a value of 0F(hex) (=15 decimal). If it has then six is subtracted from the Tens/Units register (which has the effect of decrementing the upper four bits (Tens) and setting the lower four bits (Units) to nine thus ensuring that the count down proceeds correctly.

In the UP mode, the counters are reset to zero when the Run switch is pressed at the beginning of the timing cycle and each time the routine is executed, the Tens/Units counter is incremented. In this case the Units counter will "overflow" to 0A(hex) which is ten in decimal so that six must be added to the counter to increment the top four bits (Tens) and reset the lower four (Units) to zero.

Once this has been done, the counter is checked to see that it has not been incremented to 60 and if it has, the Tens/Units counter is reset to 00 and the Hundreds incremented. The counters are then compared to the Set registers and depending on this, the program proceeds to the Timing Complete or Continue Timing section.

### ONE STEP AT A TIME

In a similar way the other parts of the program are built up until they can be written down using the instructions supported by the PIC. The two other major routines which are required are Keyboard Scan and Display which are both written

as subroutines. These are simply programs which the main program "calls" as they are required

The Keyboard Scan routine scans the switches S1 to S5 and determines which, if any have been pressed. After a short delay, the keys are read again and if the key is still pressed, its function such as Start/Stop or Set Minutes is carried out. This is done to prevent noise on the switch lines or switch bounce from being interpreted by the program as multiple key operations.

The Display routine deals with driving the display l.e.d.s, switching the various output lines as appropriate and also decoding the "counters" to provide a seven segment output on port B. A third subroutine which generates a delay is also called from within this routine. To make these subroutines "self contained", the various I/O lines which may be used as inputs in one and outputs in the other are redefined when the routines are called in the same way as in the initialise routine.

### INSTRUCTIONS

Assuming that the flowchart has been designed carefully the process of converting it to PIC instructions is quite easy as it simply involves looking up the instructions in the data sheet. These can then be converted to a machine code listing consisting of lists of 1s and 0s or Hex characters (0 to F) and programmed into the EPROM.

Because this is very boring to do and very error prone, programs (called Assembler programs) are available which do this and run on an ordinary PC. The program, written using the instruction mnemonics listed in the data sheet, together with labels to mark places to which the program has to jump and comments which help the programmer to understand what is being done, is typed into the computer using a standard word processor package and then the Assembler program is run which converts this into a format suitable for loading into the device EPROM.

Suitable PIC programmers and assemblers are now becoming available at prices which are within reach of at least the more serious hobbyist (around £100). These only require connection to one or other of the ports on an IBM PC or compatible so that devices can be programmed without having to purchase expensive programs. In addition, simulator programs are also available for use on PCs which allow the user to single step or run a program up to pre-defined break points and observe how the various registers which are displayed on the screen change as each instruction is executed.

These enable quite complex programs to be developed and debugged without resorting to the purchase or hire of a development system costing thousands of pounds. This, however, is still a high price to pay if all you want is to build a timer, so a ready programmed chip for this project is also available (See Shoptalk).

### CIRCUIT DETAILS

Now that we have looked at programming the PIC microcontroller, we shall now put it to work with a design for a Versatile Microcontrolled 3-Digit Timer. The full circuit diagram of the Timer is shown in Fig. 6. Note that the relay contacts must be rated to handle any appliance being controlled.

Transistor TR2 together with Zener diode D11 supplies +4V to the microcontroller IC1. The crystal X4 is a

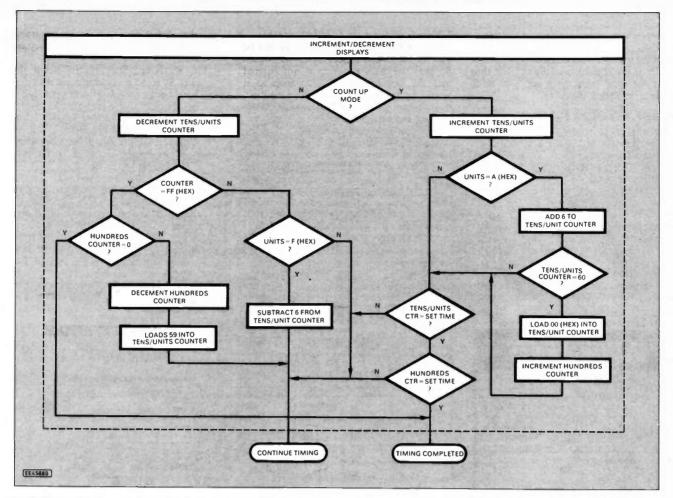


Fig. 5. Timer Up/Down count flowchart. Depending on Up or Down mode selected, the right or left hand side will be followed.

3.2768MHz type which together with capacitors Cl and C2 form the on-chip oscillator.

Time setting and other control functions are accomplished by pushbutton switches S1 to S5 and the diodes D1 to D5 are included to prevent the segment driver lines being shorted if two buttons are depressed simultaneously. The l.e.d. 7-segment displays are driven via the current limiting resistors R3 to R9, with X2 and X3 being common cathode displays while X1 and l.e.d.s D12 to D15 being connected in the common anode configuration.

The relay RLA and buzzer WD1 are switched at the appropriate times by transistors TR1 and TR3 driven from IC1 port A0 and A1 (pin 17, 18) respectively. If a relay with an operating voltage of other than 12V is used, then the circuit should be operated at the relay voltage but this should be in the range of 6V to 15V.

Any d.c. supply of the type available in the "mains plug boxes" used for powering small appliances with a suitable output voltage may be used or alternatively one can be made using a small mains transformer with a 9V-0V-9V secondary winding as shown in Fig. 7.

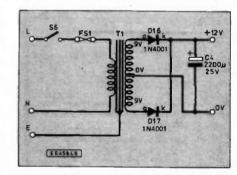


Fig. 7. Suggested mains power supply circuit.

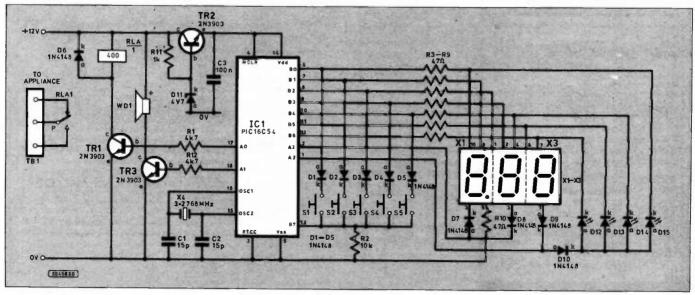


Fig. 6. Complete circuit diagram for the Versatile Microcontrolled 3-Digit Timer. Note that the display pins 1, 2, 4, 6, 7, 9 and 10 are paralled to the indentical pin on each display (X1 to X3).

### CONSTRUCTION

Construction should not present any problems if the circuit is assembled on the recommended printed circuit board (p.c.b.). This board is available from the EPE PCB Service, code 933. The topside p.c.b. component layout and full size underside copper foil master pattern are shown in Fig. 8.

Begin construction by inserting the low profile components such as resistors and diodes first, paying particular attention to the orientation of the latter. When these are in place, the taller components including the relay, buzzer and terminal block can be mounted. The buzzer is a 12V piezo type with an oscillator circuit built in and care should be taken to ensure that it is wired into the circuit the correct way around. The microcontroller ICl should not be inserted at this stage but only the 18-pin socket should be mounted on the p.c.b.

### COMPONENTS

Resistors

4k7 (2 off) R1, R12 10k R3 to R10 47 Ω(8 off)

1k All 0.25W 5% carbon film

Capacitors

C1, C2 15p ceramic (2 off) **C3** 100n polyester

### Semiconductors

D1 to D10

1N4148 signal diode

(10 off) 4V7 Zener diode

D11 D12 to 5mm red l.e.d. (4 off)

D15 TR1 to

2N3903 npn transistor

(3 off) PIC16C54 pre-programmed IC1

microcontroller (see

Shoptalk) X1 0.5in, red 7-segment high brightness l.e.d. display,

common anode X2, X3 0.5in. red 7-segment l.e.d.

display, common cathode (2 off)

### Miscellaneous

S1 to S5 s.p.s.t. push-to-make, p.c.b. mounting switches

(5 off) 3·2768MHz crystal

12V 400Ω relay, with 3A changeover contacts or RLA

WD1 Piezoelectric buzzer, with internal oscillator

Printed circuit board available from the EPE PCB Service, code 933; 18pin d.i.l. socket; 40-pin d.i.l. socket; 3-way p.c.b. mounting screw terminal block; red filter for displays; multistrand connecting wire; solder etc.

### OPTIONAL COMPONENTS (P.S.U.)

Miniature mains transformer, with

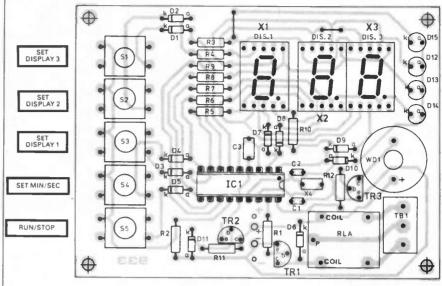
9V-0V-9V secondary winding

C4 2,200µ electrolytic capacitor, 25V D16

D17 1N4001 rectifier diode (2 off)

Approx cost quidance only

### **VERSATILE MICROCONTROLLED 3-DIGIT TIMER**



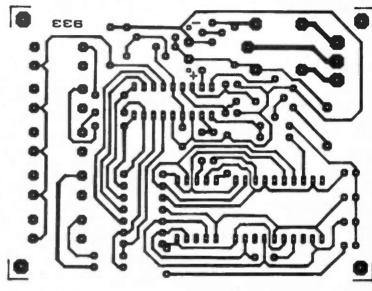
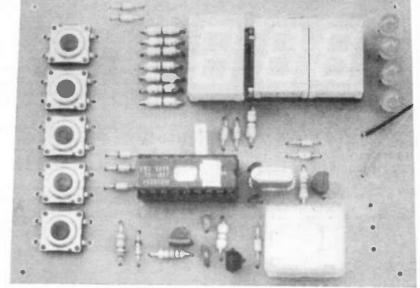


Fig. 8. Printed circuit board component layout and full size copper foil master pattern



Layout of components on the completed p.c.b.

The 7-segment displays can be soldered direct to the p.c.b. or mounted on a 32-pin socket. This is not a standard size but it can be easily cut from a 40-pin i.c. socket.

If you are using a socket, remember to plug in the most significant common anode digit as far as possible to the left in the socket and the other two common cathode digits as far as possible to the right, leaving a gap between X1 and X2. To enable the socket to be mounted, the sixth pin from the left in the top and bottom rows must be cut off or bent clear because there are no corresponding holes in the p.c.b. Remember also to fit the wire links just above the Displays.

Note that the transistors specified in the parts list have a different pinout from the more common BC548 or BC182 types. If you decide to use these, they will need to be inserted into the p.c.b. the other way around. The l.e.d. displays should also be chosen with care if the relative brightness of the digits is important.

Since the output ports of the PIC can sink more current than they can source, the common cathode units will appear brighter than the common anode display. The common anode display X1 should therefore be a high brightness type with standard units purchased for X2 and X3.

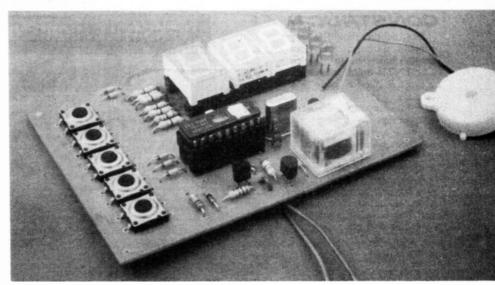
Depending on the use to which your timer is to be put and the type of box it is to be mounted in, you may decide to use panel mounted pushbutton switches in place of the ones shown in the prototype model. If this is the case, these switches will need to be connected to the p.c.b. with short lengths of wire.

When you have completed the construction, connect the 12V supply to the circuit and check that the voltage between pins 5 and 14 of the i.c. socket is no more than 4V. If all is well, switch off the supply and insert the PIC chip into its socket.

### TESTING

Since the circuit is crystal controlled there is no setting up to do and the circuit should work first time, providing, of course, there are no dry joints or solder bridges between the tracks and all the components have been fitted correctly. It should power up in the stopped mode (i.e. not counting) with the displays at zero, the relay not energised and the Mins/Secs l.e.d. on. In this mode, pressing the Start/Stop button will have no effect because no time has been set.

Press switch S1 and each time the units of seconds display should increment by one so that after ten presses it should read zero again. S3 should operate in the same way



Finished Timer board with the buzzer wired to the board. The 3-way relay contact terminal block is not shown on the p.c.b., but is mounted next to the relay.

on the minutes display and S2 on the tens of seconds display except that in this case this digit will return to zero after the sixth press enabling only 0 to 5 to be set. Note that none of the digits overflow into the next one in this mode.

Once a time has been set, pressing the Start/Stop button will cause the RUN l.e.d. to flash at 1Hz, the relay RLA to operate and the display to count down. In this mode all switches except the Run/Stop switch are disabled and pressing any of these will have no effect.

If the Run/Stop switch is pressed again, the relay will de-energise and the count will stop so that the display can be altered. If it is not altered, and the Run/Stop button is pressed again, the relay will pull in and counting will resume until zero is reached when the relay will de-energise and the buzzer will sound five times.

The l.e.d. display also flashes five times and reverts to displaying the time that was set originally, ready to do another timing run if required. The switches are also enabled so that the time can be altered if required.

In the Stop mode, pressing switch S4 will cause the Hrs/Mins I.e.d. to light and the Mins/Secs I.e.d. to go out so that although the RUN I.e.d. will continue to flash at IHz when the Start/Stop button is pressed again, timing will now be in hours and minutes and the least significant digit X3 will now change after 60 flashes of the RUN I.e.d. This will enable intervals of up to 9 hours 59 minutes to be timed.

Switch off the supply and then switch it on again but this time keeping switch S5 (the Run/Stop button) pressed. The circuit will now power up in the same state as before but this time the UP mode l.e.d. will also be lit.

The circuit will operate in exactly the same way as before except that now, if a time has been set and the Run/Stop button is pressed, the display will change to zero and count up until the preset time is reached. At the end of the timing period, when the relay de-energises and the buzzer sounds, the displays will continue to display set time (i.e. the time reached) until this is changed by operating switches \$1, \$2 or \$3 or a new timing sequence started by pressing \$55.

As noted previously, the relay has changeover contacts so that an external appliance can be switched on or switched off during the timed period depending on which output it is connected to. If required, an l.e.d. in series with a one kilohm resistor may be connected in parallel with the relay coil (l.e.d. cathode (k) to collector of TR1) to indicate the relay status.

Any other 12V relay with a coil resistance of 120 ohms or more could be used and should be chosen to have contact ratings suitable for the device being switched. Depending on the type available, it is unlikely that it will mount directly on the printed circuit board so some other method of fixing it to the box will need to be devised and the coil connected to the appropriate pads on the p.c.b.

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# Techniques

# ACTUALLY DOING IT!

### by Robert Penfold —

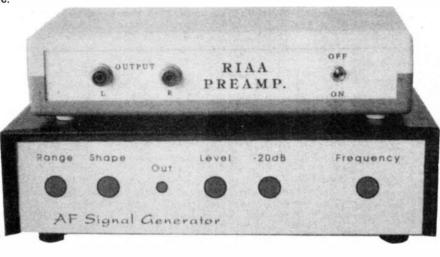
ne or two Actually Doing It articles in the dim and distant past were devoted to the subject of making high quality front panel overlays using a simple photographic process. This method was not particularly cheap, but it could produce some high quality and extremely durable results.

The articles concerned certainly provoked a fair amount of reader interest. Many constructors are clearly prepared to put some time and money into making their projects look as professional as possible.

Unfortunately, the photographic materials for making front panel overlays seem have become unobtainable. Certainly, nothing of this type seems to appear or be listed in any of the current electronic component and equipment catalogues. Computer generated panel overlays are the probable reason for the demise of the photographic methods of panel making.

good for this sort of thing. So-called "paint" type drawing programs are less than ideal, since they work at relatively low resolutions, and there can be difficulties in printing out accurately to scale.

The two types of software that are most suitable are CAD (computer aided design) and illustration programs. The differences between the two has tended to narrow over the years, but CAD programs are primarily intended for technical drawings, such as house plans, circuit diagrams, etc.



Two units labelled using a CAD drawing package.

### **RIGHT PROGRAM**

There are special programs now available for producing front panel designs, and various gadgets that can convert these into high quality panel overlays. There are also dedicated units for producing panel overlays and (or) labels electronically. Unfortunately, the cost of this sort of thing is well beyond the limits of most electronics hobbyists.

As mentioned at the end of the previous Actually Doing It, there is a cheaper alternative for those who have access to a reasonably powerful computer and a decent printer. This is a fairly involved topic, and it is one that will be considered in more detail in this month's article.

There are numerous programs that can be used to produce simple labels. Producing complete panel designs, which must be accurately printed to scale, is another matter, and requires suitable graphics software.

Specialised graphics software, such as charting and printed circuit design programs, are not likely to be much

Illustration programs are intended for general illustration work, and are in some ways less rigid than CAD programs. For example, they usually allow individual letters in words of text to be moved and manipulated in various ways, and they have sophisticated freehand drawing capabilities.

Some CAD programs can only produce relatively crude text, which is clearly a major limitation in the current context. If a CAD program will produce text of usable quality, and you only require relatively simple and straightforward panel designs, then it should provide a very quick and easy way of producing them. In general though, an illustration program is the better choice for this application.

Although programs of this kind were once extremely expensive, cut-throat competition has resulted in some sophisticated illustration software being offered at quite low prices. Programs such as "Corel Draw 3!" and "Designworks" for the PCs are reasonably cheap, and there are some

very low cost alternatives available from the usual PD/shareware sources.

Unfortunately, illustration programs mostly require a fairly powerful computer. Any reasonably modern PC, Macintosh, etc. should be able to run programs of this type, but with older PCs, and eight-bit computers it could be difficult to find suitable software.

### LAYOUT

While it is possible to simply sit down at the computer and draw up a front panel layout, this approach is not recommended. It is all too easy to draw up a very plausible looking layout that turns out to be completely impractical when you actually start fitting things in place.

The safer approach is to work out a panel layout by placing control knobs onto the front panel of the case and working out a layout that is both attractive and practical. The fixing nuts of toggle switches, jack sockets, etc. can be used to represent these components, and help you to visualise the finished product. Small pieces of "Bostik Blu-Tak" can be used to hold things in place, plus it will also give you the

freedom to alter the layout.

Remember there are two sides to every panel. Just because everything fits nicely and is quite workable from the front side, it does not mean that there will be sufficient space for everything on the reverse side. Measure the components carefully to make sure that your layout has reserved enough space for them.

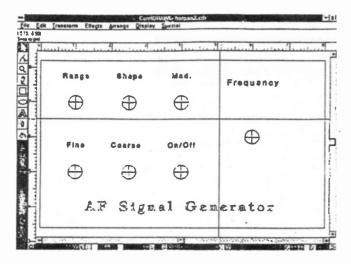
Also, bear in mind that most cases have internal obstructions. For example, built-in

mounting pillars for circuit boards, and the screws, etc. that hold the two halves of the case together. Some obstructions, such as unwanted mounting pillars, can be removed. Others are essential parts of the case, and you have to work around them.

Drawing to scale with an illustration or CAD program should be very straightforward. There is usually provision for a grid of on-screen dots to provide accurate reference points, plus a co-ordinate display showing the current cursor position.

Some programs have rulers marked across the top of the screen and up one side. These normally have moving markers which respond to changes in the position of the cursor. The cursor can be made to "snap" precisely to grid points, so that accurate drawings are assured.

Start by drawing a rectangle to represent the perimeter of the front panel. The position of each control and socket must be carefully measured on the "dummy" layout, and transferred to



Screen dump from "Corel Draw 2!" The user interface is a standard Windows type. Note the on-screen rulers.

the drawing as accurately as possible. Lines, circles, etc. can be drawn, and it is quite easy to draw up a template that can be taped in place on the panel, and used as a guide when drilling the holes.

Most programs have layers, or some other method of letting you print out only the required elements of the drawing. In most cases the drilling template and the actual panel overlay can therefore be incorporated into a single drawing. This method ensures that there are no embarrassing discrepancies between the two!

### RIGHT TYPE

Illustration software is usually supplied with a reasonable range of fonts (lettering styles), with many more available as optional extras. These days a virtually unlimited range of lettering sizes can be selected. For normal labels it is probably best not to get carried away, and to select a fairly simple font ("Swiss", "Times", "Helvetica", etc.). Many programs allow the spacing

Many programs allow the spacing between letters to be varied. The default spacing is usually very small, which does not give the most readable labels. Increasing the spacing slightly usually gives more comprehensible results.

Obviously some of more zany fonts might be applicable to projects of the "fun" variety, such as electronic games. Also, it is common practice to use a less formal font for name labels ("Signal Generator", "Stereo Preamp", etc.). This is a matter of using a little common sense.

When adding panel labels by hand it can be difficult to get them accurately centred above their controls and sockets. There is no problem of this type with computer produced panels, since the program should have a facility to automatically centre each piece of text above a given point.

Virtually all illustration programs provide proportional spacing, and in many cases kerning as well. Kerning is simply where pairs of letters such as "AW" and "PA", which have complementary shapes, are moved slightly closer together. Otherwise they tend to be perceived as being too far apart.

It may be possible for the user to manually adjust the position of

each letter. This is probably not worth the efwith fort the smaller legends, but it might be worthwhile with large labels. It is when using large letters that your perception of perfect kerning is most likely to be at odds with the program designer's opinions on the subject.

With illustration programs (and some CAD types) there are various "fill options" that can be used on

any part of the panel. It is therefore possible to have grey rather than black letters, or to have a pale grey background instead of a white background. In practice this type of thing is unlikely to produce particularly good results unless you have access to a very high resolution printer.

Using various colour fills is a very different matter. If you have access to a colour printer it is likely to be well worthwhile experimenting with a few colour schemes. Once again though, try not to get carried away, and try to choose colour schemes that are appropriate to each project.

### **ICONS**

It is increasingly common for readymade equipment to have controls, etc. marked with icons instead of normal labels. For instance, a volume control could be marked with a simple wedge shaped icon rather than being labelled "Volume"

This method is a useful one where panel space is limited, and it is difficult to accommodate normal labels. It does not require a great deal of artistic talent in order to draw up a few simple icons.

### **HARD COPY**

With most printers it is only possible to produce the hard copy on some form of paper. Coloured paper is available for most printers, and can help to produce some attractive panels.

It is advisable to use a fairly heavy paper, because the lighter types are practically translucent, and do not usually provide very attractive results. Also, they tend to go slightly blotchy when glued in place. Coloured ribbons/ink cartridges are available for some printers, enabling the labels to be printed in something other than a "Ford" choice of colours.

The problem with a paper overlay is that with use it soon becomes dirty and discoloured. A covering of self-adhesive transparent plastic film is virtually essential.

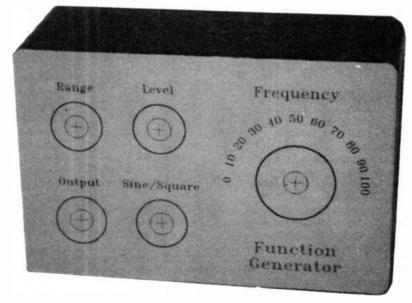
With larger panels it can be difficult to get a wrinkle and blister free covering. Experience has taught, that it is best to remove the backing from the plastic film, and then lay it in place, starting at one corner and gradually working across to the opposite corner.

If necessary, with some films it is possible to peel the material back slightly and try again without damaging the overlay. Fortunately, if it should all go wrong it does not cost much to print out a new overlay and try again.

Laser printers are not restricted to use with paper, and they can be used with a variety of plastic films, including some self-adhesive types. These can be used to produce some very professional looking panels, but they tend to be very expensive.

Matters are made worse by the fact that these films are often only sold in packs of about 100 or 200 sheets. Some suppliers offer sample packs for little or nothing, and these provide an opportunity to experiment a little at low cost.

Modern graphics software is much easier to use than the equivalent software of a few years ago. It still takes some hours of effort to really get to grips with a good illustration or CAD package, but it should be well worth the effort.



It is useful to use a CAD package which can rotate legends around controls.

# VIDEOS ON ELECTRONICS

A range of videos designed to provide instruction on electronics theory. Each video gives a sound introduction and grounding in a specialised area of the subject. The tapes make learning both easier and more enjoyable than pure textbook or magazine study. They should prove particularly useful in schools, colleges, training departments and electronics clubs as well as to general hobbyists and those following distance learning courses etc.

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capacitors, etc are used in common circuits.

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### **VCR MAINTENANCE**

VT102 84 minutes: Introduction to VCR Repair. Warning, not for the beginner. Through the use of block diagrams this video will take you through the various circuits found in the NTSC VHS system. You will follow the signal from the input to the audio/video heads then from the heads back to the output.

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Each video uses a mixture of animated current flow in circuits plus text, plus cartoon instruction etc., and a very full commentary to get the points across. The tapes are imported by us and originate from VCR Educational Products Co, an American supplier.

(All videos are to the UK PAL standard on VHS tapes)

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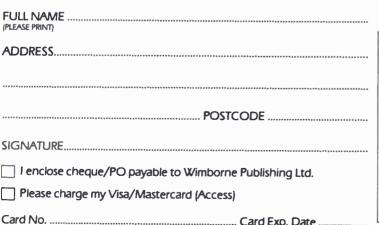
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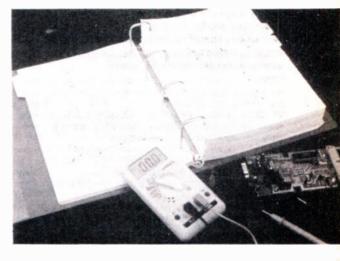
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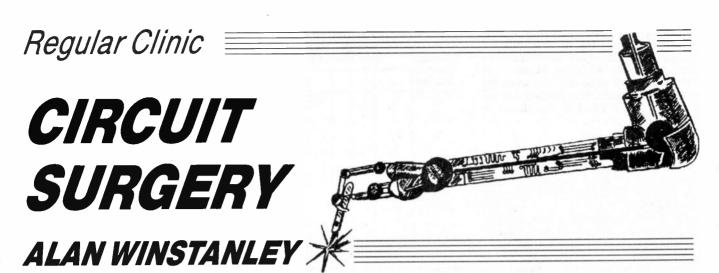
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Our regular column handling readers' questions and problems looks at a precision rectifier with no forward voltage drop, also a peek at BCD codes plus a single transistor pulse generator. Last but not least, some queries answered by our Internet feed!

A salways, a wide variety of questions and queries come our way hoping for a helping hand, so here's a selection which we hope will be of general interest to readers. Let's start with a question from *Mr. K.E. Langford* from Sidmouth in Devon, who asks:

### **Active Precision Diode**

I recently encountered a problem which I've not seen discussed in EPE, perhaps you could help. What's the best circuit (for amateur hobbyists) to make a precision rectifier which eliminates the normal diode drop?

In my simple Windicator constructional project (last month) for example, I used a germanium diode to rectify the motor voltage output. This suffers from a 0·2V drop which consequently clips the signal at low wind speeds/motor voltages, though I thought it would not be a problem in that simple project. An application such as this could perhaps use a "precision rectifier" circuit, one which has the unidirectional behaviour of an ordinary diode but without an undesirable voltage drop, so none of the signal will be lost. (Silicon diodes drop 0·6V or so when conducting.)

A standard active rectifier which is based around an op.amp IC1 is shown in Fig. 1. Essentially it's a non-inverting amplifier circuit with a diode in the feedback path. This forms a half-wave

rectifier which will clip any negative-going transitions at the input. During positive input transitions, the op.amp "follows" the input signal and the cathode of D1 is fed back to the inverting input of the op.amp, providing negative feedback, which sees a lower voltage returned than that at the +VE input. Consequently, IC1 compensates for the diode drop by swinging higher to lose the difference in voltages arising between the two inputs.

For negative transitions, the diode clips the signal. The result is a very accurate rectifier which has no forward voltage drop, and which behaves almost like an ideal half-wave rectifier. Its accuracy will be maintained even at low millivolt levels. For best performance, choose an op.amp with a good "slew rate" (SR) – the parameter which describes how quickly the op.amp can actually respond by swinging its output voltage.

Measured in volts per microsecond (V/μs), if you have specific requirements then you will select an op.amp with a high SR figure. For example: the bipolar 741C has an SR of (only) 0.5V/μs – slow! – whilst at the other extreme a device such as the AD845K can change at up to 100V/μs – pretty fast! Clearly the speed of response of the op.amp determines how quickly it can follow the input signal, but in many hobbyist

applications I doubt if the SR value would be too critical. I've also drawn in a second op amp IC2 which acts as an optional buffer for the rectified signal, preventing any excess loading of the precision rectifier output.

### **UJT** Oscillator

My thanks to Tony Dacro of Morpeth for reviving memories of the once very common "unijunction transistor" oscillator. Tony posed the following question:

I'm assembling a trigger circuit for operating a silicon controlled rectifier. The trigger needs to operate at up to 20,000 times per second. Could you suggest a simple circuit based around the 2N2646 I'm interested in using?

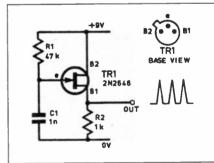


Fig. 2. Unijunction transistor oscillator.

Ah, yes ... the 2N2646! The unijunction transistor (UJT) has largely fallen into disuse, it would seem, though it was very popular in our project pages in the mid-late 1970s. It's possible to build a simple one-transistor circuit which generates a sawtooth-type waveform which could be used for firing triacs and thyristors, or could operate at audio frequencies as a simple audio oscillator.

A circuit, with values chosen to run at roughly 20kHz, is shown in Fig. 2. TR1 is a UJT and it conducts between emitter (e) and b1 when the RC network of R1/C1 has charged to a critical trigger voltage. The capacitor discharges through the b1 resistor resulting in a spike output, and the process repeats.

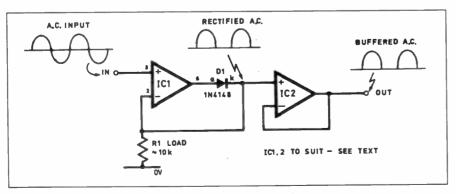


Fig. 1. Precision rectifier based around two op.amps.

The frequency can be altered by making R1 variable, say 10k to 100k; alternatively the capacitor can be increased in value to reduce the output frequency. It should operate easily with values in the low microfarards. The 2N2646 has an unusual pinout which is also shown in the circuit schematic.

### **BCD Switches**

All is not what it appears, at least not if you buy a rotary switch which actually has a "BCD output". *Richard Beesley* of Tonbridge, Kent asks for some advice.

I recently acquired a switch which is a ten-position thumbwheel type (though instead of a thumbwheel it has two pushbuttons to advance or retard the switch setting). It has a common terminal and four others. As a beginner in electronics, I'd appreciate it if you could unscramble the switch for me – in particular, can I use it to produce ten outputs?

It sounds like you have a switch which has a "BCD" output. Binary Coded Decimal (BCD) is a simple method of expressing a decimal digit using a four-bit code. BCD switches rely on the pattern of the copper tracks within the switch's printed circuit board mechanism to generate four binary digits – "ons" and "offs" if you like – which are the binary equivalents of the numbers I to 10 shown on the switch button. This is sometimes required for interfacing in certain logic applications. It works as shown in Table I.

If you apply a Logic 1 to the "common" input and dial in a decimal number on the switch, a binary coded decimal number will be outputted at the four output terminals. For example, the table shows how the decimal number 7 is represented by the BCD code 0111 (which is decimal 0 + 4 + 2 + 1) which appears at the four output pins.

Although you have a BCD-version of the switch, Richard, you could convert the BCD code to decimal using a TTL decoder chip. The 74LS42 is a 1-to-10 BCD decoder/driver. Simply input the BCD code generated by the thumbwheel into pins 12 to 15 and out pops the decimal equivalent to emulate a one-pole 10-way switch. Fig. 3 shows the pin-outs of the 74LS42.

I could not see this chip advertised in the Maplin catalogue but Electromail and several others do list it. The similar 74LS45 is a high current version with open collector outputs. These can be pulled up with resistors up to +30V d.c.;

pin 16

+5V 1 2 4 8 9 8 7

BCD Input

74LS42

Decimal 0 - 9

0 1 2 3 4 5 6 GND

pin 1

Fig. 3. Pinouts for the 74LS42 BCD decoder/driver chip.

Table 1.BCD switch logic.

Decimal Switch	BCD Switch Output			
Setting	8	4	2	1
0	0	0	0	0
1	0	0	0	1
2	0	0	1	0
3	0	0	1	1
4	0	1	0	0
5	0	1	0	1
6	0	1	1	0
7	0	1	1	1
8	1	0	0	0
9	1	0	0	1

the output pins can sink up to 80mA when low. Alternatively you could perhaps have acquired a one-pole ten-way switch instead, Richard!

### NiCad Discharging – More!

My column in Circuit Surgery, February 1995 included a simple design for a nickel-cadmium (NiCad) discharger. This prompted a few comments from *Chris Flynn* of Beverley, North Humberside.

I was a little concerned about the circuit given for a NiCad Discharger. One of the requirements for a sound design is that it does not take the voltage of the battery pack down to a level which risks one of the cells running down to zero volts, or worse, eventually reversing its own polarity!

A safe rule of thumb is to cut off discharge down to "Nominal Voltage – IV". For example, a 6V pack would be taken down to 5V, an 8·4V pack down to 7·4V, and so on. This is manageable up to 10 cell (= 12V) battery packs. Also, whilst writing, I appreciated the return of Ingenuity Unlimited – always the first thing I used to read in PE!

Thanks for writing, Chris, and point taken. In fact the basic design I published was only intended as a blow-out-proof load for a NiCad, having current limit and reverse voltage features, together with a token measure of automatic cut-off in that the LM317 regulator will in any case cease conduction when its input voltage falls below roughly 3V d.c. or so. It made no reference to the actual time spent discharging, of course.

I have actually performed quite some research on this topic, since I found

conflicting "rules of thumb" at the time and definite word from anyone particular. There lively were discussions on this topic on the Internet recently, especially in the mostly American sci.electronics newsgroup. The consensus of opinion seems to be that there is no

such thing as "memory effect" in NiCad rechargeable cells. In particular, there is a need to distinguish between a cell's capacity and its terminal voltage. It is said to be beneficial to discharge NiCads down to 25 per cent of their capacity every once in a while. My information additionally is that it's a good idea to discharge down to only 10 per cent of stated capacity occasionally, or instead use a light load only to discharge to a terminal voltage of 1V per cell.

The chemistry behind this is complex. The idea is to prevent larger cadmium crystals from forming within the cell. Smaller "trickle" discharges help to form a small uniform microcrystalline structure which optimises the discharge performance of the cell. Larger crystals are more difficult to dissolve during the discharge process, eventually causing a build-up of internal resistance. This causes voltage depression, which commonly goes under the misnomer of "memory effect".

Trawling the Internet in tireless search of interesting stuff, I did come across an on-line document available by ftp (file transfer protocol) to anyone who's got an Internet connection. This document is a twelve page summary of various news items taken from the electronics newsgroups and is extremely interesting to read, with contributions from Hewlett Packard engineers and quotes from General Electric.

At the time of writing, the document is available by anonymous FTP from ftp.armory.com /pub/user/rstevew/nicad.faq.zip - note that the two dots in the filename may confuse DOS-based software (rather than UNIX), so it may be easier to get the file ftp'd on your behalf and emailed to you by an ftp server such as that at Imperial College. Then you can set about PKunzipping it when it's delivered!

For those readers not on the Internet who would like to review it, please send a large A4 size SAE (38p stamps UK only, or 2 IRCs) plus 60p to cover the copying – to Circuit Surgery and I'll send be happy to send you a reprint of the articles. Ask for the document "Circuit Surgery NiCad Text" – it's ten pages long and worth reading!

### Micro Lab News!!

Our Internet feed is very lively and apart from enabling overseas readers to contact us – whether about subscription rates, back issues or anything else – it also enables me to join in lively discussions in the sci.electronics newsgroup. In fact, it's too addictive! Plus, EPE does its bit, flag-waving and posting general information and comments such as book reviews to which the broadly American newsgroups seem extremely receptive. Dave Howey in Cape Town, South Africa (but soon emigrating to the U.K., he tells me) asked by email:

Hi Alan, what happened to the "Robo Spot" project which is mentioned at the end of your Teach-In No. 7 book? I'm very interested in robotics and am waiting in keen anticipation of this project < grin > . David. Howey@chameleon.alt.za

Quite a number of readers who followed our highly popular Teach-In '93

Series successfully - re-published in book form as Teach-In No. 7, available directly from our Direct Book Service (see their pages for ordering information) - have asked about the progress of the Micro Lab "add-ons" which we mentioned both at the end of the series and also in the book. In particular, we had in mind at the time of design a potential idea for an electronic "buggy" which could interface to the Micro Lab, a project we code-named "Robo Spot" in a flurry of enthusiasm!

Regrettably it has not proven possible to complete the design for this particular add-on accessory, for the following reason. Whilst the Mini Lab design (also in Teach-In No. 7) was aimed at a very wide audience, and to date has sold well over a thousand units (and rising), the Micro Lab was of necessity a much more complex design which restricted its appeal to the more advanced section of the Teach-In readership, and in practice the number of the more specialised Micro Labs which were purchased could never reflect the enormous popularity of Mini Lab sales. Therefore, it was simply not viable for the Teach-In team to input such a great deal more design work into a buggy when only a few dozen units may ultimately be purchased.

Nonetheless the excellent Micro Lab microprocessor trainer/applications unit does indeed have quite a following, and since we do want to encourage further use of the design, Micro Lab designers Keith Dye and Geoff MacDonald tell me confidently that they do intend to complete two additional aspects of the Micro Lab. So, stop press, Micro Lab fans will be pleased to hear that work is in hand with the following peripherals:

Micro Lab Parallel Port - to enable program listings to be dumped to an external printer through a new parallel port. This will require a revised version EPROM, which will also include several new commands - notably HEXDUMP and DISASSEMBLE, the latter enabling the user to print program listings in "plain English" (disassembled language) rather than unfriendly hexadecimal code. It's planned to include an Epsom printer emulator which many types of dot matrix printer (and laser printers even) should be happy with.

Micro Lab-PC Serial Link - this will consist of a small piggy-back p.c.b. which sits on the Micro Lab board, to provide a 9-pin D-type serial link which can connect with the serial connection of an IBM-compatible PC or a terminal emulator. The most notable feature is that it will enable users to read or write Micro Lab programs on their own PC, and save any program listings on hard disk! This will necessitate both an exchange EPROM and a revised (swapout) PLA chip.

Work is at an advanced stage but the final specifications are subject to confirmation. So there you have it - still plenty more to look forward to if you've invested in this highly versatile design! Naturally, Circuit Surgery will keep you posted of any further news. (Note that both Mini Lab and Micro Lab kits continue to be available from Magenta Electronics Ltd., see their advertisement elsewhere in this issue.)

In Circuit Surgery next time: what those "IP" ratings for enclosures actually mean, l.e.d. flashing circuits, disposal of etchant fluid - how? - and more from our Internet feed! Circuit Surgery is your column. which relies heavily on your input. If you have any topical questions, follow-ups or comments, or any subject which you think will interest your fellow readers which you would like me to investigate, then please write to:

Alan Winstanley, Circuit Surgery, Wimborne Publishing Ltd., Allen House, East Borough, Wimborne, Dorset, BH21 1PF, United Kingdom. Or email to alan@epemag.demon.co.uk. As always, your scribe endeavours to respond to every query but cannot guarantee to do so unless we intend to use it in this feature.

Next month: Ingenuity Unlimited, another round up of readers' own circuits. See you soon!

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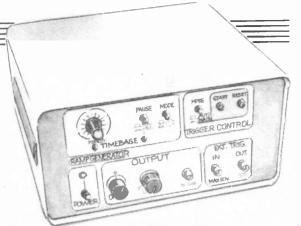
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# Constructional Project

# RAMP **GENERATOR**



# NEIL JOHNSON BEng(Hons) AMIEE Part 2

A professional quality ramp generator aimed at the intermediate to advanced constructor.

AST month it was explained how the Ramp Generator project evolved from a need to expand and improve the author's range of test equipment with a versatile unit which could be used for a wide range of tasks. Detailed descriptions of how the circuit worked were then given.

This month we conclude by describing the construction and testing of the unit.

### LOGIC BOARD **ASSEMBLY**

Details of the component layout on the double-sided Logic board are shown in Fig. 7. From hereon, this side of the p.c.b. is referred to as Side B. Side A, not surprisingly, is the reverse side!

Start construction with the small passive components – resistors R1 to R17 and capacitors C1 to C25. Continue by inserting terminal pins for l.e.d.s D1 and D2, and the two transistors TR1 and TR2. Next, fit the d.i.l. sockets.

Of the remaining items, connector PL1 can be fitted next. Rotary switch S1 needs to be fitted into its mounting bracket, then the whole assembly soldered to the board, including the bracket tags.

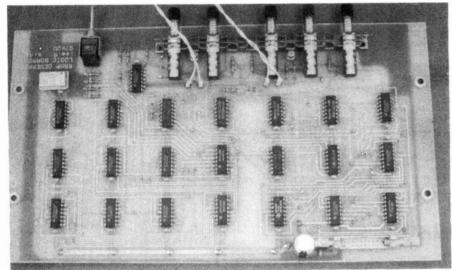
Before mounting, switches S3 and S4 need to be made momentary action. This is accomplished by carefully removing the small leaf spring and latching pin from each of them. Then fit the array of switches, S2 to S6, into their mounting bracket, and fold over the fixing tabs. Carefully fit the button caps to the switches and solder the complete assembly to the logic board.

The two l.e.d.s, D1 and D2, are mounted on the front panel, connected to the logic board via flying leads - a length of twistedpair can be used for this. Ideally the léads need to be about 15cm to 20cm long, covering them with insulation sleeving.

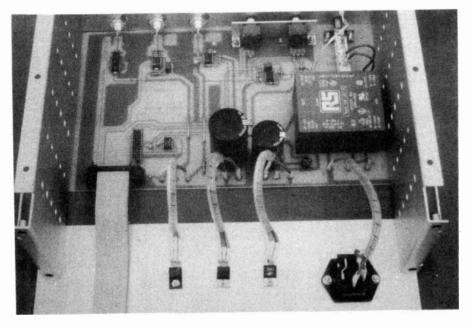
### ANALOGUE BOARD *ASSEMBLY*

Details of the component layout for the Analogue board, which is single-sided, are shown in Fig. 8.

Start off assembly by soldering in the wire links. Those shown in heavy lines should be made of fairly chunky



The assembled Ramp Generator Logic Board. Note that the switches are mounted integrally with the common brackets.



The assembled Analogue board, complete with the power supply components, mounted inside the case. Note how the voltage regulators and the mains input socket are connected to the rear panel.

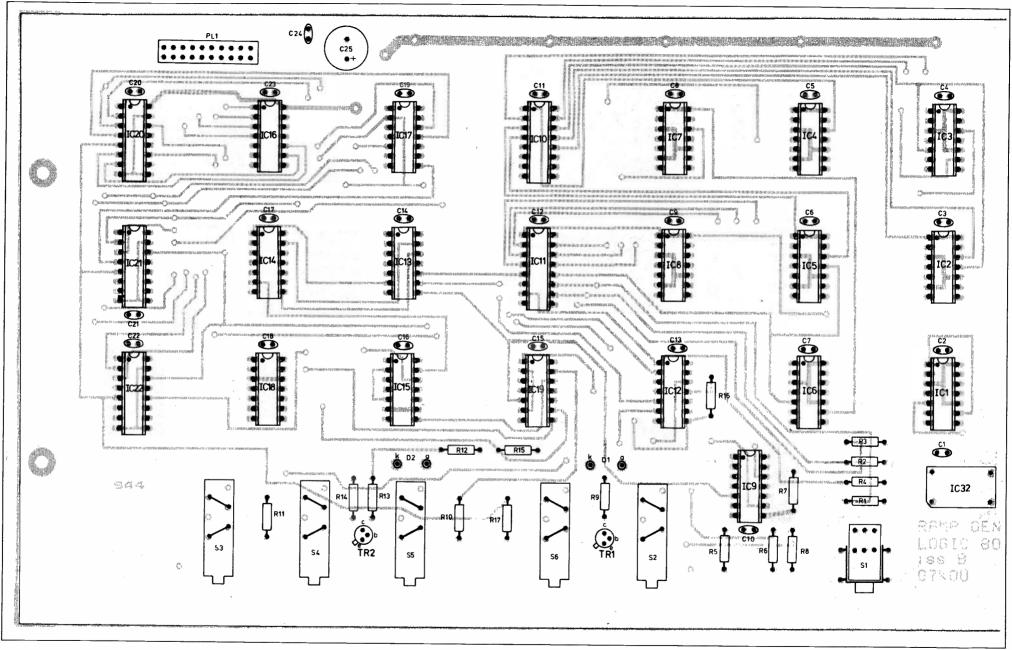


Fig. 7. Component layout details for the Logic board of the Ramp Generator. Only the upper side tracking is shown. The horizontal extremes of the board are not shown – they are vacant of components but include the mounting holes.

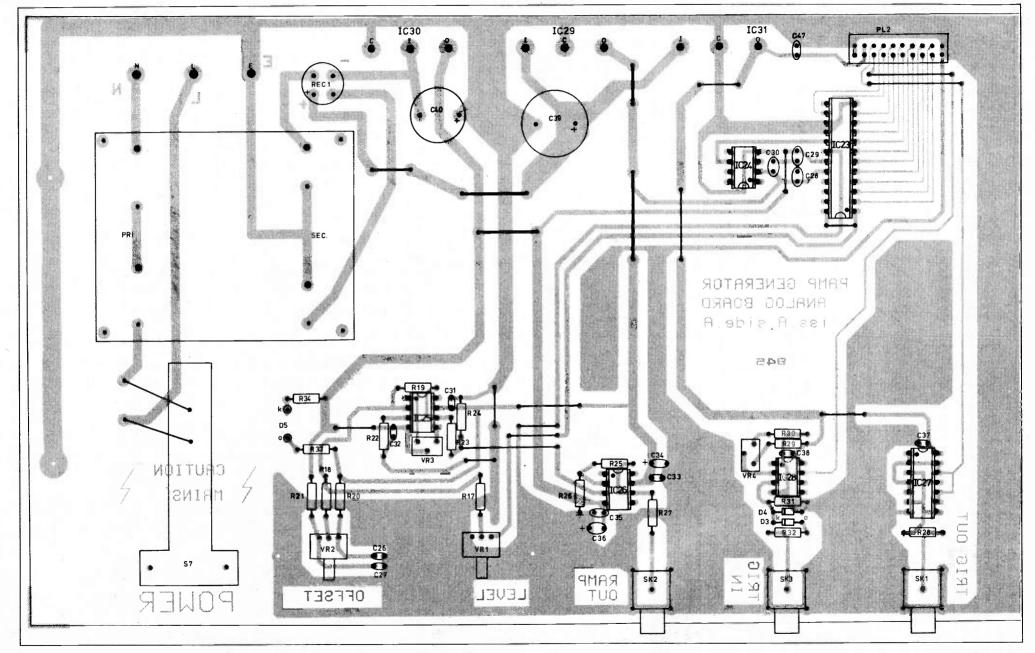


Fig. 8. Component layout details for the Analogue board of the Ramp Generator. This board is single-sided. The horizontal extremes of the board are not shown – they are vacant of components but include the mounting holes.

tinned copper wire, such as 18s.w.g. The remainder can be 22s.w.g. or 24s.w.g. tinned copper wire. There are nineteen wire links to be made. Follow on with terminal pins for all off-board connections, including those for l.e.d. D5 and the short links to switch S7.

All passive components in order of size—resistors, capacitors, tantalum capacitors, presets, electrolytic capacitors—are soldered to the board, taking care to correctly fit the polarised capacitors. The larger items, connector PL2, transformer T1 and switch S7, are fitted next. Transformer T1 has an asymmetric pin layout so it should be impossible to fit it the wrong way round.

Switch S7 is fitted into its mounting bracket, which is then bolted to the board. Use insulated wire links to connect the switch to the board.

With the semiconductors, start with diodes D3 and D4, and then the bridge rectifier REC1, making sure you get them the right way round! Follow with the three voltage regulators, IC29 to IC31, which are connected to the board via flexible wire leads. These should be made of 16/0·2 or 19/0·2 stranded wire about 15cm long. The 100n decoupling capacitors C41 to C46 are soldered directly to the regulator leads on the back of the board (Side A).

Next fit d.i.l. sockets for the integrated circuits. Finally, fit the mains connector SK4 to the board with flying leads. These should be made of stranded wire, capable of carrying about 3A, about 15cm long with both soldered ends covered in heatshrink tubing. Connect l.e.d. D5 to the board in the same way as D1 and D2.

The last stage of construction of the analogue board is to mount the two potentiometers VR1 and VR2. A metal bracket supports them on the board, thus avoiding fixing them directly to the front panel. The bracket details are shown in Fig. 9. Fix the potentiometers to the bracket before bolting the assembly to the board.

### CASE IT!

Assuming that the recommended case is being used, the front panel has four mounting studs welded to it. Drill out the required 15 holes in the front panel, as shown in Fig. 10, and five holes in the rear panel, as shown in Fig. 11, ensuring that they align with board-mounted components where necessary. Make sure the holes are cleanly finished and free from any burrs.

Clean both front and rear panels with white spirit or methylated spirits before applying lettering and protective varnish. The front panel legends of the prototype were made by photocopying the design

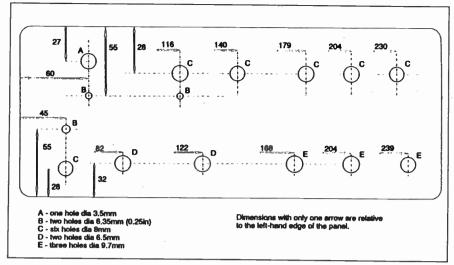


Fig. 10. Front panel drilling details.

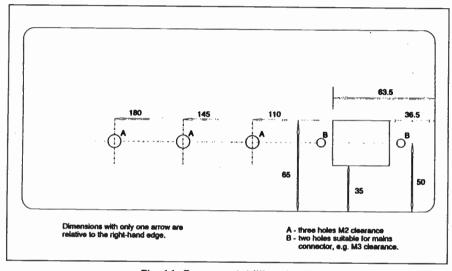


Fig. 11. Rear panel drilling details.

(seen in the photos and available as previously commented in Part One) onto adhesive-backed paper. This was then covered with sticky-backed plastic before the individual sections were cut out and stuck on to the panel.

Next fit the remaining panel hardware: l.e.d. clips and BNC sockets SK1 to SK3. Bolt the front and rear panels to the two side panels, using the nuts and bolts supplied, to make a sturdy frame onto which the boards are to be attached.

Starting with the Analogue board, bolt this to the underside of the frame, components facing upwards, carefully sliding the shafts of the two potentiometers and power switch through their holes in the front panel. Then attach the mains connector and voltage regulators to the rear panel using M3 and M2 screws respectively. Do not forget the mica washers and silicone heatsink grease for the regulators.

Complete the Analogue board by pushing l.e.d. D5 into its front panel clip and connecting the BNC sockets (SK1 to SK3) to the board with short lengths of tinned copper wire.

Before fitting the Logic board, a cable link is required between connectors PL1 and PL2. This should be made from about 15cm of 20-way IDC ribbon cable and two 20-way IDC connectors.

Connect the cable link into PL1 on the Logic board, then flip the board over so that the component side is facing downwards and the switches are facing the front. Rest the board on the top of the frame while you fit the l.e.d.s D1 and D2 into their clips. Slide the board forward so that the switches protrude through their front panel holes, finally bolting the board firmly to the frame.

Now for the last fiddly bit. Turn the unit over and carefully push the other end of the cable link into connector PL2 until it clicks into place. Finally, attach the three knobs to the shafts of panel potentiometers VR1, VR2, and switch S1.

Before fitting the bottom half of the case, screw on its tilting feet. Then carefully fit

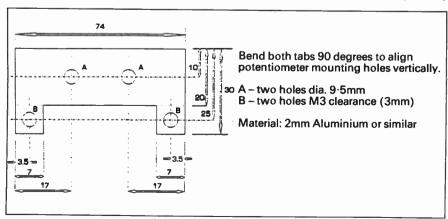
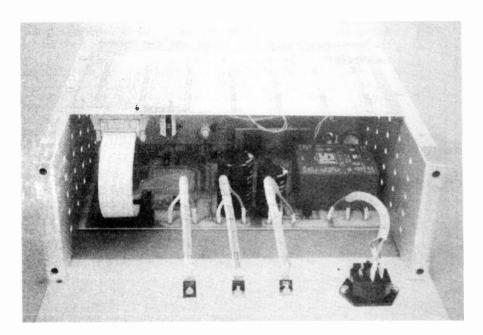


Fig. 9. Details of the mounting bracket for potentiometers VR1 and VR2.



The Logic board is mounted on top of the case frame, upside down.

the central frame into the shell before fitting the top shell. The two halves are attached to the frame with the dark-brown coloured side bars. Fit these into place and secure them with the matching screws – a penny makes an ideal screwdriver for these large headed screws.

# TESTING... TESTING...

DO NOT PLUG THE UNIT INTO THE MAINS YET.

To perform the following tests you will need an Ohmmeter, reading up to several hundred kilohms, and an Oscilloscope. Begin by checking the mains power inlet.

There should be a very low resistance (a few milliohms) between the Earth pin and the case, with about 120 ohms between the Live and Neutral pins with the Power switch in the ON position. There should be a virtual open circuit between Live or Neutral and Earth – check this with the ohmmeter on its highest resistance range, or better still with a Mega-ohmmeter which tests at about 500 volts.

The following tests will involve operating the Ramp Generator with part of the cover removed. Because of the presence of mains voltages, PRECAUTIONS SHOULD BE TAKEN:

- Do not make adjustments with the unit connected to the mains. Disconnect it BEFORE you make any adjustments, then reconnect and test.
- When the mains is on, keep one hand in your pocket – if you should be careless, at least you will not provide the mains current with a path across your chest.
- Place the Ramp Generator on a wooden or plastic surface, clean off any solder splashes, swarf, strands of wire or any other conducting material.

To make adjustments to the two presets VR3 and VR4, the front panel must be removed.

Connect a mains lead to the mains input connector and an oscilloscope to the Ramp Output connector SK2. Set the Ramp Generator Period switch SI to 100ms, Pause switch S2 off, Auto triggering switch S5 on, Offset potentiometer VR2 set midway for zero offset, and Output Level control VRI set for about 2V.

Turn on the power. The power l.e.d. D5 and Sweeping l.e.d. D2 should come on and a ramp waveform be visible on the scope screen (the scope's timebase and input controls may need adjusting).

If all is well, check the operation of the Period, Pause and Ramp Mode controls (S1, S2, S6). Then check that the Level and Offset controls work correctly. If fitted, the Offset Null preset VR3 can be trimmed so that the middle point of the Offset Null control VR2 relates to an offset of zero.

The last section to check is the Trigger Control. Set the Trigger Mode (S5) to Manual and check that the output goes to zero (plus any offset). Operating the Start button (S4) should produce a single ramp. On slower period settings the Reset button (S3) should instantly stop the ramp and reset the unit. Also check that the Sweeping l.e.d. D2 remains on during the ramp sweep time and turns off at the end.

The final adjustment is the External Trigger Input Threshold, VR4. This can be set with the aid of a variable power supply and an additional push-to-make switch. Set the supply to the threshold level you want, say 2V, and use the pushswitch to provide a pulse of this voltage to the External Trigger

Input SK3. Then adjust VR4 until the Ramp Generator just triggers on the voltage pulse.

When you have completed the testing stage replace the front panel and any other panels you have removed. You are now the proud owner of a superb piece of bench test equipment with many applications. Now then, how about some...

### **APPLICATIONS**

Apart from looking nice with lots of buttons to press and knobs to twiddle, the Ramp Generator has a wide range of uses.

The first use to which the prototype was put was to check the timebase of the author's Oscilloscope. The period of the ramp waveform, being crystal-controlled, is accurate to within ±0.01%; most analogue scopes use an R-C timing circuit accurate to a couple of per cent.

To check the accuracy of the scope timebase select a convenient period, say Ims, and set the scope timebase to ten times this value, in this case 10ms. If the scope is roughly in calibration, a complete ramp waveform should be seen, from start to finish, covering the complete 10cm horizontal graticule. If not, get out the scope's service manual!

If your main interest is electronic music you could use the Ramp Generator to control an oscillator, filter or amplifier. This would create single or multiple sweeps of pitch, cut-off or amplitude with a great degree of control.

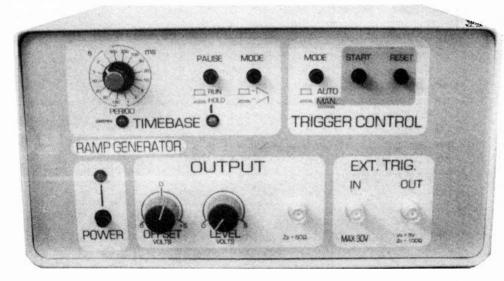
If you have an X-Y plotter lying around, the Ramp Generator could provide it with a precision timebase. In fact, an X-Y plotter, or oscilloscope in X-Y mode, combined with the Ramp Generator can form the basis of a wide range of test equipment. A few suggestions are listed here, plus example block diagrams shown in Fig. 12.

### **Curve Tracer**

The simple Curve Tracer circuit in Fig. 12a can plot a  $V_{ce}$  versus  $I_c$  curve for an *npn* transistor for a particular base current  $I_b$ .

### Frequency Response Analyser

The Frequency Response Analyser circuit shown in Fig. 12b will produce a plot of gain or attenuation for a network over a particular range of frequencies; ideal for filter designers. Or the network could be replaced by a loudspeaker and



Full size photocopies of the front panel legends are available free from the Editorial office.

(a) CURVE TRACER

RAMP IN

(b) FREQUENCY RESPONSE ANALYSER

X AXIS FREQUENCY

NETWORK

VCO

NETWORK

VCO

Y AXIS
FREQUENCY

Fig. 12. Ramp Generator application examples.

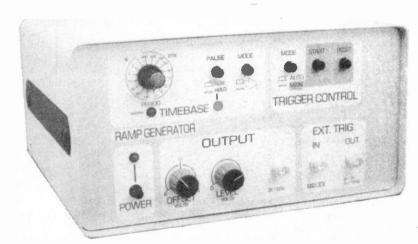
microphone/amplifier pointing at each other to plot the frequency response of a loudspeaker system.

### Spectrum Analyser

How would you like a Spectrum Analyser? As suggested in Fig. 12c, all that is needed is a voltage controlled filter (VCF – e.g. SSM2044) with a bandpass output, and a detector diode. This would work well at audio frequencies, or for an RF Spectrum Analyser you would need an RF VCO (voltage controlled oscillator), mixer, low-pass filter and detector diode (if you are not sure of what the circuit would look like you probably wouldn't understand how it would work – it's beyond the scope of this article to explain it!).

### ON THE UP

The Ramp Generator is an accurate and versatile piece of equipment with a wide range of uses. Experienced constructors will recognise that there are many other uses beyond those outlined here. It is hoped that this project will generate much interest amongst other readers, to design and build a wide variety of accessories – probably even come up with ideas the author has never thought of. Perhaps even to publish their designs in *EPE*. Enjoy it!



# RING BINDERS FOR EPE

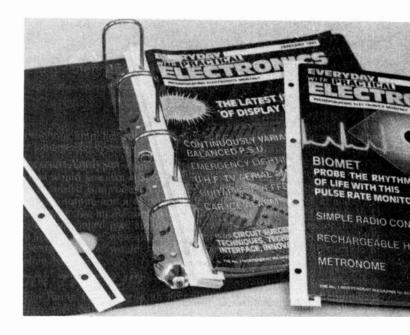
This ring binder uses a special system to allow the issues to be easily removed and reinserted without any damage. A nylon strip slips over each issue and this passes over the four rings in the binder, thus holding the magazine in place (see photo).

The binders are finished in hard wearing royal blue p.v.c. with the magazine logo in gold on the spine. They will keep your issues neat and tidy but allow you to remove them for use easily.

The price is £5.95 plus £3.50 post and packing. If you order more than one binder add £1 postage for each binder after the initial £3.50 postage charge, (for overseas readers the postage is £6.00 each to everywhere except Australia and Papua New Guinea which costs £10.50 each).

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# REPORTING

# AMATEUR RADIO

# 氏

# Tony Smith G4FAL

### **NZ PROPOSES NO-CODE**

The International Amateur Radio Union recently came down firmly on the side of keeping the amateur Morse test for the foreseeable future, but there is now a new development in this ongoing controversial debate.

In New Zealand, an organisation called ORACLE (Organisation Requesting Alternatives by Code-Less Examinations) has bypassed New Zealand's national radio society, NZART, and has approached their licensing authority (the Ministry of Commerce) direct to suggest that the NZ government propose the abolition of the amateur Morse test at the next World Radio Conference (WRC-95), in Geneva in October.

Last September, NZART received an assurance from the Minister of Commerce, that New Zealand would not actively make proposals to change the international regulations "until such time as there is evidence of significant opinion here in New Zealand, and/or overseas" to support such change.

In March NZART received a further letter announcing that the government had now decided to propose to *WRC-95* the deletion of Radio Regulation 2735 which requires a Morse test for amateur radio h.f. operation.

### WHO WILL SUPPORT IT?

At present, the various IARU membersocieties around the world are discussing with their own government delegations to *WRC-95* the implications for amateur radio of the various items on the WRC agenda.

In theory, they will now close ranks and oppose a proposal that not only goes against their present internationally agreed policy, but undermines their status as collective representatives of the radio amateur fraternity to their licensing authorities.

The question is, will they? Or will some break ranks and support the New Zealand government's proposal? The last time something like this happened was at the 1979 World Administrative Radio Conference when several different countries put forward proposals to amend RR 2735.

At that time a Morse qualification was required for operating on frequencies below 144MHz. Sweden suggested the figure be reduced to 28MHz and Papua New Guinea suggested 30MHz.

The United States government delegation put forward a proposal that if approved would have resulted in the abolition of the Morse test requirement altogether. This proposal was made independently by the US administration in the face of strong opposition from the American Radio Relay League, which had a mandate from its membership to keep the test.

The US proposal was favoured by two administrations and opposed by at least fifteen others. Finally, it was withdrawn, the proposal from Papua New Guinea was adopted, and it is that which forms the present-day Radio Regulation relating to Morse code.

### AMATEUR RADIO AND THE INTERNET

There are many facilities for radio amateurs on the Internet. This isn't the place to explain how to access them, or to include long lists of "addresses", but suffice to say there are collections of radio data, directories, organisations, news bulletins, callbooks, specialist groups, and individual radio clubs worldwide to be found by simply pressing a few buttons and making inexpensive (often local) telephone calls.

With new Internet facilities being developed all the time, I'm wondering if all those people said to be waiting to come into amateur radio, "if only the Morse test could be abolished", will continue to find the prospect of combining their interests with our hobby so attractive?

ORACLE (above), says that the number of computer enthusiasts under the age of 25 exceeds the total number of radio amateurs of all ages by many times (and this is no doubt true), and that future recruitment into amateur radio is more likely to be linked to computing interests than anything else. But will it?

The Internet can already provide communication facilities in many ways better than amateur radio. There is no compulsory technical examination, no Morse test to pass, and no need to set up an expensive radio station and antenna over and above the cost of a computer. There are no problems through causing radio interference to the neighbours, and there are no regulations restricting what you can talk about with other parties.

There is even software available now which provides real-time two-way voice or videophone function. One system, Internet Phone, developed in Israel, provides two-way simplex contacts, in effect just like amateur radio.

This includes a chat mode where you list your name under a particular topic. Then anyone with the same interest (and software) can click on your name for a one-to-one conversation (QSO!).

### **RADIO IS THE COMMON INTEREST**

Why bother with amateur radio at all when such superior facilities are now available elsewhere? Well, there is one other factor which I omitted to mention. Radio itself.

Amateurs are interested in more than just a communications link, the radio function itself is an important part of the hobby. They talk about many

non-technical matters, but they always get back to their "rigs", their antennas, their ancillary equipment, discussing their performance or their problems.

They talk about propagation, low power achievements, DXing, and many other radio related matters. Furthermore, they enjoy the challenge offered by the vagaries of radio propagation, and the uncertainties of what will happen each time they go on the air.

That's why I suggest computer enthusiasts would find the Internet more satisfying. There are plenty of radio amateurs who enjoy using their computers on the air, but who are also interested in the radio side of things, and I don't mean them. But if amateur radio is simply to be a substitute for the telephone line, that surely can't be the way forward for a hobby that still claims to be scientifically and experimentally orientated?

As can be seen from CRACLE's remarks, most of the fuss about the amateur Morse test is related to the computer boom. Could it be that the communications explosion via the Internet will take the pressure off and we can all settle down again, enjoying our own separate interests within amateur radio, and still be linked by our mutual interest in the medium itself?

### **NEWS FROM THE RSGB**

The Radio Society of Great Britain has published a new book aimed primarily at Novice licensees. *Practical Transmitters for Novices* contains a selection of easy-to-build transmitter designs suitable for the UK Novice bands. It includes designs for 160m double-sideband speech (DSB), 80m CW (Morse), and 6m FM speech transmitters.

This is one of a number of books specially written for Novices which are available from RSGB Sales (Telephone 01707 659015 for further details). All RSGB books are available to nonmembers but cost a litle more than to members.

If you want to see these and other RSGB publications before buying, the RSGB will have book and information stands at the following events in the next two months.

August: 6th, RSGB National Mobile Rally, Woburn Abbey; 19/20th, RSGB Stafford Amateur Radio and Computer Show; 27th, Torbay Amateur Radio Society Annual Mobile Rally.

September: 3rd, Telford Rally; 10th, Southend and District ARS Radio and Computer Rally. Phone the RSGB (number as above) for details of venues, etc.

The RSGB's National Convention will be at the Stafford Show, and will include a lecture programme covering packet radio, repeaters, the Novice licence, contesting, and learning Morse.

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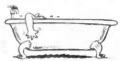
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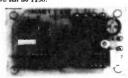
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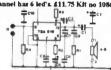
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OMP MOS-FET POWER AMPLIFIERS

THOUSANDS PURCHASED BY PROFESSIONAL USERS



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VISA

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OMP/MF 200 Mos-Fet Output power 200 watts OMP/MF 200 Mos-ret Output power 200 Watts
R.M.S. into 4 ohms, frequency response 1Hz - 100KHz
-3dB, Damping Factor >300, Slew Rate 50V/uS,
T.H.D. typical 0.001%, Input Sensitivity 500mV, S.N.R.
-110 dB. Size 300 x 155 x 100mm.
PRICE \$64.35 + \$4.00 P&P

OMP/MF 300 Mos-Fet Output power 300 watts R.M.S. into 4 ohms, frequency response 1Hz - 100KHz - 3dB, Damping Factor > 300, Slew Rate 60V/uS, T.H.D. typical 0.001%, Input Sensitivity 500mV, S.N.R. -110 dB. Size 330 x 175 x 100mm. PRICE £81.75 + £5.00 P&P

OMP/MF 450 Mos-Fet Output power 450 watts R.M.S. into 4 ohms, frequency response 1Hz - 100KHz - 3dB, Damping Factor > 300, Slew Rate 75V/uS, T.H.D. typical 0.001%, Input Sensitivity 500mV, S.N.R. -110 dB, Fan Cooled, D.C. Loudspeaker Protection, 2 Second Anti-Thump Delay. Size 385 x 210 x 105mm. PRICE £132.85 + £5.00 P&P

OMP/MF 1000 Mos-Fet Output power 1000 watts R.M.S. into 2 ohms, 725 watts R.M.S. into 4 ohms, frequency response 1Hz - 100KHz -3dB, Damping Factor > 300, Slew Rate 75V/uS, T.H.D. typical 0.002%, Input Sensitivity 500mV, S.N.R. -110 dB, Fan Cooled, D.C. Loudspeaker Protection, 2 Second Anti-Thump Delay. Size 422 x 300 x 125mm. PRICE \$259.00 + \$12.00 P&P

NOTE: MOS-FET MOOULES ARE AVAILABLE IN TWO VERSIONS: STANDARD - INPUT SENS 500mV, BAND WIDTH 100KHZ. PEC (PROFESSIONAL EQUIPMENT COMPATIBLE) - INPUT SENS 775mV, BAND WIDTH 50KHZ. ORDER STANDARD OR PEC.

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PRICE \$32.71 +

£2.00 P&P

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PRICE 122.71 + £2.00 P&P
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RES. FREO. 71Hz, FREO. RESP. TO 7KHz, SENS97d8.

PRICE 23.74 + £2.50 P&P
10" 200 WATT R.M.S. ME10-200 GUITAR, KEYB'D, DISCO, VOCAL, EXCELLENT HIGH POWER MID.
RES. FREO. 65Hz, FREO. RESP. TO 3.5KHz, SENS 99d8.

PRICE £43.74 + £2.50 P&P
12" 100 WATT R.M.S. ME12-100LE GEN. PURPOSE, LEAD GUITAR, DISCO, STAGE MONITOR.
RES. FREO. 49Hz, FREO. RESP. TO 6KHz, SENS 100d8.

PRICE £35.64 + £3.50 P&P
12" 100 WATT R.M.S. ME12-100LT (TWIN CONE) WIDE RESPONSE, P.A., VOCAL, STAGE
MONITOR. RES. FREO 42Hz, FREO. RESP. TO 10KHz, SENS 98d8.

PRICE £36.64 + £3.50 P&P
12" 200 WATT R.M.S. ME12-200 GEN. PURPOSE, GUITAR, DISCO, VOCAL, EXCELLENT MID.
RES. FREO. 58Hz, FREO. RESP. TO 6KHz, SENS 98d8.

PRICE £46.71 + £3.50 P&P
12" 300 WATT R.M.S ME12-300GP HIGH POWER BASS, LEAD GUITAR, KEYBOARD, DISCO ETC.
RES. FREO. 47Hz, FREO. RESP. TO 5KHz, SENS 103d8.

PRICE £26.71 + £3.50 P&P
15" 200 WATT R.M.S. ME15-200 GEN. PURPOSE BASS, INCLUDING BASS GUITAR.
RES. FREO. 46Hz, FREO. RESP. TO 5KHZ, SENS 99d8.

PRICE £270.34 + £4.00 P&P
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RES. FREO. 39Hz, FREO. RESP. TO 3KHZ, SENS 99d8.

PRICE £270.34 + £4.00 P&P

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RES. FREC. 40Hz, FREC. RESP. TO 5KHz, SENS. 9948.
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RES. FREC. 35Hz, FREC. RESP. TO 3KHz, SENS 9648.
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8" 60WATT EB8-60TC (TWIN CONE) HI-FI, MILTI-ARRAY DISCO ETC.
RES. FREQ. 40Hz, FREQ. RESP. TO 18KHz, SENS 98dB.
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