# Electonic Engineering

JUNE 1953



Y AMPLIFIER FEATURES Critically damped : Rise Time 0.15 µ.s. Bandwidth D.C .--- 3.5 Mc/s : Sensitivity 20 mv/cm to 70V/cm : **Amplitude Gating eliminates Overload Defects: Differential Input Circuits** 



WAVEFORM MONITOR

OSCILLOSCOPE TYPE WM 3 E.M.I. FACTORIES, HAYES, MIDDLESEX

ADDITIONAL FACILITIES **Direct Time Measure**ment to  $\pm 2\%$  F.S.D. from 0.5 µ.s. to 40 ms. Sync./Trig. Amplifier with Amplitude-Frequency-Phase Discrimination : Direct Voltage Measurement A.C./D.C. to  $\pm 2\%$  F.S.D. from ±20mv to ±500V.



TWO SHILLINGS

## Polytetrafluoroethylene a new material for material for FLECTRONICS

During the past few years a new resin-like substance, Polytetrafluoroethylene has been developed for Commercial use. It is unique among organic compounds in its chemical inertness, in its toughness over a wide range of temperatures, and in its low dielectric losses over a wide range of frequencies.

P.T.F.E. has zero water absorption with low surface tension, high impact strength and form stability, a far greater resistance to chemicals than either Gold or Platinum, and retains its strength and dielectric properties at temperatures ranging from minus 100° C to plus 250° C. Its co-efficient of friction is of a very low order.

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P.T.F.E. has been successfully used in a wide range of highest grade typeapproved valveholders made by The Edison Swan Electric Co., Ltd., and a range of lead-through and stand-off insulators is available. Also a number of stock sizes of Sheet, Rod, Tape, Yarn, Slugs etc., and moulded or fabricated parts can be supplied to specification.

Its arc resistance, heat resistance and low electrical losses suggest unlimited applications within the Electronic Industry.

For full technical details please apply to

## THE EDISON SWAN ELECTRIC CO. LTD.

Sales Department P.T.F.E.4, 21 Bruton Street, London, W.1. TELEPHONE: MAYFAIR 5543.

> Head Office: 155 Charing Cross Road, London, W.C.2 Member of the A.E.I. Group of Companies.

E.R.22

## CLASSIFIED ANNOUNCEMENTS

The charge or these advertisements at the LINE RATE (if under 1' er 12 lines) is : Three lines or under 7/6, each additional line 2/6. (The line averages seven words.) Box number 2/- extra, except in the case of advertisements in "Situations Wanted," when it is added free of charge. At the INCH RATE (if over 1' or 12 lines) the charge is 30/- per inch, single column. Prospectuses and Company's Financial Reports 2/4 05. 0d. per column. A remittance must accompany the advertisement. Replies to box numbers should be addressed to : "Electronic Engineering," 28, Essex Street, Strand, London, W.C.2. Advertisements must be received before the 14th of the month for insertion in the following issue,

#### OFFICIAL APPOINTMENTS

OFFICIAL APPOINTMENTS ASSISTANTS SCIENTIFIC'S The form provides down insistioners invite applications for provides down in the applications for provide down in the applications of the forces, but applied to the completion as a whole or in one of provide down in the applications for the applications of down in the applications for the applications of the provide down in the applications of the forces, course provide down in the applications of the forces, course provide down in the application forms form provides of the forces, course (i) Biological Sciences (i) General work ranging over provides of the forces, course (i) General (i) and highly biological Sciences (i) General (i) and highly provides (i) application forms form provide downse, o) application forms form provide downse, o) application forms form provide downse, of the force of the force sciences (i) provide downse, of the force of the forces, course provide downse, or ference and application forms form provide downse, of the forms and minetened formations provide downse, or ference and application forms form provide downse down application form forms form provide downse down application forms form provide downse down

Buildon, W. 1 (Enclosing addressed foolscap w 2447 **BR IT IS H ELECTRICITY AUTHORITY.** London Division. General Assistant Engineer (Communications). Technical Department. Duties in connexion with construction, testing and maintenance of Authority's Private Com-munications System, including equipment using Line, Teleprinter, and Radio Techniques. Pre-vious experience of this work essential experience in telephone manufacturer's Works valuable additional qualifications. Preference given to applicants having University Degree with First or Second Class Honours in Engineer-ing or Physics. Salary and conditions in accord-ance with NJ.B. Agreement. Schedule C within Class AX, Grade 8, Salary 2526 to 2670 19s. rising to ultimate maximum £749 14s. p.a. Application may be made on form obtain-able from Divisional Secretary, B.E.A., London Division, Generation House, Gt. Portland Street, W.1, and be received within 14 days of this advertisement.

Street, W.1, and be received within 14 days of this advertisement. W 2430 CROWN AGENTS FOR THE COLONIES. Assistant Signals Officer required by the Govern-ment of Nigeria for the Aviation Department for one tour of 18 to 24 months in the first instance. Salary, etc., either (a) in scale £750 rising to £1,315 a year with prospect of per-manency or (b) in scale £807 rising to £1,433 a year, on a temporary basis with gratuity at the rate of £100 a year. Outfit allowance £60. Free passages for officer and wife, and assist-ance towards cost of children's passages, or their maintenance in the United Kingdom. Liberal leave on full salary. Candidates must have a knowledge of the fundamental principles of electricity and magnetism and of radio en-gineering with experience in the maintenance of aeronautical radio transmitters and receivers. direction finders, test equipment and small petrol and diesel engine generator sets. Work-shop experience and a knowledge of radar will be of advantage. Minimum examination qualifications are C. & G. certificate in radio communications or technical electricity, or satisfactory pass in M.C.A. radio mechanics course. Apply in writing to the Crown Agents,

4 Millbank, London, S.W.1, stating age, name in block letters, full qualifications and experi-ence and quoting M2C/29637/EK. W 2496

JUNIOR TECHNICIAN wanted. Opportunity to train in high voltage X-ray and physics laboratory work while studying. Salary accord-ing to age and experience. Applications to the Director, Radiotherapeutic Research Unit. Ham-mersmith Hospital, W.12. W 2477

MINISTRY OF SUPPLY requires Physicists or Engineers for R.A.E., Farnborough, Hants, to work either on problems connected with generation and distribution of electric power in aircraft, or on problems of electric power in aircraft, some knowledge of mathematics desirable. Qualifications: Higher School Certifi-cate (Science) or equivalent but further training in Physics or Electrical or Mechanical Engineer-ing to standard of Degree or H.N.C. may be an advantage. Salary within ranges: Experi-mental Officer (minimum age 26), £597-£754, or Assistant E.O., £264-£555. Women some-what less. Posts unestablished. Application forms from M.L.N.S., Technical and Scientific Register (K), 26, King Street, London, S.W.1, quoting D.146/53A. Closing date 12th June, 1953. W 2434 MINISTRY OF SUPPLY requires design Engin-

Register (K), 20. King Street, London, S.W.I., quoting D.146/53A. Closing date 12th June, 1953. W 2434 MINISTRY OF SUPPLY requires design Engin-cers at Harwell to assist on design problems of control and instrumentation for new types of reactors. Qualifications: British of British parents. Recognized electrical apprenticeship or equivalent training instrumentation or elec-tronics and A.M.I.C.E., Mech.E. or E.E. or exempting qualifications. Experience in design and installation of instrumentation and control systems and servo mechanisms. Salary and Grade according to age. qualifications: experi-ence within £597 (Eng. III age 25)-£1,454 (Eng. I maximum). Housing accommodation available later if married. Not established, but oppor-tunities to compete for established. Street, London, S.W.1, quoting (D 186/53A). Closing date 13th June, 1953. W 2464 MINISTRY OF SUPPLY requires Electrical Engineer at Harwell to assist design of new control systems, inspect during manufacture, supervise installation; or maintain and develop evisiting nualifications. Experience in large-scale remote recording and control installations (temperature flow, PH, etc.), remote power control servo-mechanism, ionization instru-ments, high stability amplifiers, ratemeters, scelers and discriminators. Salary, £597 (age 25)-£928. Not established, but opportunities to compete for established, but opportunities to compete for established, but opportunities to compete for establishement may arise. Applica-tion forms from M.L.N.S., Technical and Scientific Register (K). 26. King Street, London, S.W.1, quoting D.185/53A. Closing date 12th June, 1953. W 2478

June, 1953. W 2478 NEWCASTLE UPON TYNE Education Com-mittee. Rutherford College of Technology. Principal: Dr. H. A. Scarborough, M.A. Appli-cations are invited for the post of Assistant Grade B in the Department of Electrical Engin-ering. Applicants should possess graduate or equivalent qualifications. Ability to teach Tele-communications subjects an advantage. Salary in accordance with the Burnham Technical Scale for Grade B Assistants, £490-£765 with additions for approved qualifications and train-ing. Application forms which may be obtained from the undersigned should be returned within 14 days of the appearance of this advertisement. H. V. Lightfoot, Director of Education, City Education Office. Northumberland Road, New-castle upon Tyne, 1. W 1714 POST OFFICE: Assistant Engineer. The Civil

POST OFFICE: Assistant Engineer. The Civil Service Commissioners announce an open com-petition commencing 18th August, 1953, for not less than 33 posts for men. Age 174-234 on 1st March, 1953, with extension for regular service in H.M. Forces, and up to two years for other Forces service or for established Civil Service. For particulars (including prospects) and application form (preferably by postcard) to Secretary, Civil Service Commission, Burling-

ton Gardens, London, W.1, quoting No. 159/53. Application forms must be returned by 25th June, 1953. W 2483

Application forms must be returned by 25th June, 1953. W 2483 **PROFESSIONAL ENGINEERS** in various Government Departments. The Civil Service Commissioners invite applications for about 150 pensionable posts in a wide variety of mechani-cal and electrical engineering duties. Applica-tions may be accepted up to 30th September, 1953, but an earlier closing date may be announced. Candidates must be under 35 on 1st January, 1953, with extension for regular service in H.M. Forces and up to two years for permanent Civil Service. For the Post Office they must be at least 21, for the Ministry of Supply and Ministry of Civil Aviation at least 23, and for all other Depart-ments at least 23, and for all other Depart-ments at least 23 and for all other Depart-ments at least 23, and for all other Depart-ments at least 24, in or exemption from, Sections A and B of the corresponding Asso-ciate Membership examinations. Exceptionally, candidates of high professional attainment, but without the specified qualifications, may be admitted. For some posts Associate Fellowship of the Royal Aeronautical Society or an Honours Degree in Physics will be accepted. Inclusive salary scale (men in London) £628 (at age 21) to £970. Starting salary according to age up to £875 at 34. Candidates entering below. age 25 will start at salaries varying from £429 at age 21 to £549 at age 24. Prospects of pro-functions Salaries of next higher grades are £970-£1,280 and £1,331-£1,536. Somewhat lower for women and in the Provinces. Further particulars and application forms from Civil Service Com-mission, Scientific Branch, Trinidad House, Old Burlington Street, London, W.1, quoting No. S85/53. W 2473 THE COLLEGE OF AERONAUTICS. Appli-cations are invitted for the following posts in

Burlington Street, London, W.1, quoting No. S85/53. W 2473 THE COLLEGE OF AERONAUTICS. Appli-cations are invited for the following posts in expanding Electrical Section: (1) Lecturer in Electrical Engineering (Electronics), whe will be required to cover fundamental electronic cir-cuits, including pulse circuits and their applica-tion to the measurement of non-electrical quantities. He will also assist in the development of the specialized electronic instruments required by all Departments of the College in their research work. Honours Degree in Physics or Electrical Engineering required with experience in at least one of the following fields: airborne radio and radar systems; servomechanisms; electronic techniques in measurement and con-trol. Salary within range £600.£1.000 p.a, with superannuation under F.S.S.U. and family allowance. Further particulars available. (2) Design Draughtsman, who will be responsible for the layout and detail design of the great variety of electronic and electrical equipment developed by the Section. Applicants must have had experience with electronic and/or light electronic and applicants must have had experience and addresses of not more than three referees, to the Recorder, The College of Aeronautics, Cranfield, Bletchley, Bucks. W 2465

Bucks. W 2465 UNIVERSITY COLLEGE of the Gold Coast. Department of Physics. Applications are in-vited for the appointment of an Electronics Technician in the Science Workshops. Some experience of instrument making is required. The appointment will be on a five-year contract in the first instance. Salary £800 p.a. plus a gratuity of £100 per annum in lieu of superan-nuation provision. Part furnished residential accommodation at rent of not more than 10 per cent of salary will be provided. Family allowances of £50 per child (up to a maximum of £150) if children are resident in the Gold Coast or £100 per child (up to a maximum of £300) if children are resident elsewhere. Free first-class passages will be provided for the selected candidate and his family on assumption and relinquishment of the appointment and for annual leave of eight weeks in the United Kingdom. Outfit and baggage allowances are paid on first appointment. The duties will

#### OFFICIAL APPOINTMENTS (Cont'd.)

include the construction, maintenance and repair of electrical equipment, the training of staff and responsibility for electronic com-ponents stores. Applications (six copies) with the names of three referees, should be sent to the Assistant Registrar (London), University College of the Gold Coast, 29, Tavistock Square, London, W.C.1, from whom further particulars may be obtained. Closing date: 30th June, 1953. W 1710

June, 1953. W 1710 UNIVERSITY OF ABERDEEN. Electronics Mechanic required for Physics Laboratory, Natural Philosophy Department; without super-vision, able to construct complex apparatus from theoretical circuit and design simple circuit to specification; some general workshop experi-ence useful; wage 2365 to 4450, according to experience. Applications to the Secretary, University of Aberdeen. W 1711

University of Aberdeen. W 1711 VACANCIES in Government Department. Leading Draughtsmen and Draughtsmen are required, with experience in the layout of tele-communication and electronic equipment, in-volving detailed mechanical design, preparation of all mechanical drawings, sub-assemblies, and final assembly, circuits, specifications, and stock-lists; suitable for prototype and batch pro-duction manufacture. Practical workshop experience, and knowledge of modern methods an asset but not essential. Salaries (Consoli-dated Provincial Rates). Leading Draughtsman, £592 0s. 0d. x £20 0s. 0d. to £702 0s. 0d. Draughtsmen, £374 0s. 0d. x £20 0s. 0d. to £597 0s. 0d. Apply in writing: Personnel Department, G.C.H.Q., 53 Clarence Street, Cheltenham. W 2495

#### SITUATIONS VACANT

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The engagement of persons answering these advertisements must be made through a Local Office of the Ministry of Labour or a Scheduled Employment Agency if the applicant is a man aged 18-64 inclusive or a woman aged 18-59 inclusive unless he or she, or the employment, is excepted from the provisions of the Notifica-tion of Vacancies Order, 1952.

A DESIGN DRAUGHTSMAN with experience of layout of prototype electronic equipment and test gear is required by the Research Labora-tories of The General Electric Co. Ltd., North Wembley, Middlesex. Candidates should also have some experience of designing electro-mechanical mechanisms. Apply in writing to the Staff Manager (Ref. RLO/4), giving full details of age, qualifications and experience. W 2472

W 2472 A DESIGNER is required to organize and control a Section engaged on the design and specification of electrical and electronic equip-ment intended for production testing in a valve and cathode ray tube factory. This work involves familiarity with a wide range of tech-niques including pulse R.F., E.H.T., ci-cuits and automatic switching systems. It is desirable that he shall have Graduate or equivalent status and essential that he can claim considerable practical experience in this or closely allied fields. Write, giving full particulars of experi-ence, etc., to Box No. W 2451. A DESIGNER/DRAUGHTSMAN with Hinber

A DESIGNER/DRAUGHTSMAN with Higher National Certificate is required by the Research Laboratories of The General Electric Co. Ltd. North Wembley, Middlesex, for work on micro-wave receiving equipment. Some previous experi-ence in this field is desirable, but not essential. Apply in writing to the Staff Manager (Ref. RLO/763), stating age, qualifications and experi-ence. W 2419

A HIGH VACUUM Impregnation Unit-or R.I.C. Specification 214 or individual require-ments. Good delivery at competitive prices from Vactite Components & Assemblies. 505, Lord-ship Lane, S.E.22. Forest Hill 7089. W 166 AERO RESEARCH LTD. (A Ciba Company) wish to appoint a salesman, with special know-ledge of the technology of electrical insulation, to promote the sales of Araldite (epoxy) cast-ing and laminating resins. This is a new position in an expanding industry. Applications should be made in writing, giving are and full details of previous experience, to Managing Director, Aero Research Ltd., Duxford, Cambridge. W 2456

AIRTECH LIMITED. Aylesbury & Thame Airport, Haddenham, Bucks. require Radio Mechanics and Testers, capable of working on own initiative to Ministry specifications. Small batch production. Varied and interesting work & Thame

covering Audio to U.H.F. Preference given to applicants with experience on Aircraft Radar installation. Write giving age, details of ex-perience and qualifications. W 1729 AN INSTRUMENT ENGINEER required in Research Laboratories for circuit design of electronic and electrical measuring apparatus. Applicants must have three years' design ex-perience of electronic apparatus and minimum qualification of Higher National Certificate. Progressive post for successful applicant. Salary according to age, experience and qualifications. Full details to Personnel Officer, Ericsson Tele-phones Limited, Beeston, Nottingham. W 2437 AN INSTRUMENT MECHANIC is required phones Limited, Beeston, Nottingham. W 2437 AN INSTRUMENT MECHANIC is required for duties in the Electronics Division of the Research and Experimental Department. Appli-cants should possess a high Degree of mechani-cal skill and be able to participate intelligently in both design and manufacture of smal. electro-mechanical devices. Experience in a similar capacity, whilst desirable is not essential. Write, stating age, experience and salary expected to the Personnel Office, Saunders-Roe Limited, East Cowes, I.O.W. W 2428

APPLICATIONS ARE INVITED for the undernoted vacancies: Electronics Development Engineer with University Degree or equivalent ualifications, also practical experience in design and development work. Technical and Laboratory Assistant with Higher National Cer-tificate in Electrical Engineering or City and Guilds Telecommunication Certificate. Prefer-ence given to person with practical knowledge of radar and laboratory work. Electronics Engineer with knowledge and practical experi-ence in micro-waves, electro-mechanical measurements and servo-mechanisms. Appli-cations by letter giving details of education, ualifications and experience to The Secretary, Messrs. Barr and Stroud Limited, Anniesland. Glaagow, W.3. APPLICATIONS ARE INVITED for the

ASSISTANT ENGINEER for installation and maintenance duties X-ray and Electromedical apparatus. Training given if necessary, but sound basic knowledge electrical engineering essential. Write M.S.A. Ltd., 53, Park Royal Road, London, N.W.10. W 2474

essential. Write M.S.A. Ltu., 55, rais toom Road, London, N.W.10. W 2474 ASSISTANT ENGINEER required for general development work on magnetic amplifiers and aircraft equipment. in a company situated about thirty miles S.W. of London. Applicants should preferably possess a Degree or Higher National Certificate in Electrical Engineering. Apply to Box No. W 1719. BELLING & LEE LTD., Cambridge Arterial Road, Enfield, Middlesex, require research assistants in connexion with work on electronic components, fuses, interference suppressors and television aerials. Applicants must be graduates of the I.E.E. or possess equivalent qualifica-tions together with similar laboratory ex-perience. Salary will be commensurate with previous experience: five day week, contributory pension scheme. Applications must be detailed and concise, and will be treated as confidential. W 138 EDITICH INSULATED CALLENDER'S

and concise, and will be treated as confidential. W 138 BRITISH INSULATED CALLENDER'S CABLES LTD. require for their Telecommuni-cations Laboratory. Kirkby Trading Estate, near Liverpool, capable draughtsmen for interesting design and development work on light Electro/Mechanical accessories. Applicants qualified to O.N.C. standard and with previous experience in a similar capacity preferred. Only keen and conscientious men need apply. Appli-cations in writing to: The Staff Offieer, B.I.C.C. Ltd., Prescot, Lancs. W 2399 BRITISH INSULATED CALLENDER'S CABLES LTD. require for their Telecommuni-cations Laboratory. Kirkby Trading Estate, near Liverpool, Laboratory Mechanics for the development of light electro-mechanical acces-sories. Previous experience in polythene press moulding would be an advantage but this is not essential. Candidates for these vacancies should be first-class men capable of precise bench work. Applications in writing to: The Staff Officer, British Insulated Callender's Cables Ltd., Prescot, Lancs. W 2409 BRITISH INSULATED CALLENDER'S Ltd., Prescot, Lancs. W 2409 **B R I T I S H** INSULATED CALLENDER'S CABLES LTD. require for their Telecommuni-cations Laboratory, Kirkby Trading Estate, near Liverpool, capable draughtsmen for interesting design and development work on light electro/ mechanical accessories. Applicants qualified to O.N.C. standard and with previous experience in a similar capacity preferred. Only keen and conscientious men need apply. Applications in writing to: The Staff Officer, B.I.C.C. Ltd., Prescot, Lancs. W 2442 CHHEF DESIGNER required by Scientific CHIEF DESIGNER required by Scientific Instruments and Electro Magnetic Assemblies Manufacturer located at Stevenage, Hertfordshire. This appointment, which will arise in July, calls for a highly qualified man with administrative ability and offers an exceptional opportunity. Practical experience with Relays, Temperature Control Equipment and Magnetic Amplifiers an advantage. Pension Scheme and house available. Send details of age, qualifica-tions, experience and present salary to Box No. W 2438.

No. W 2438. CHIEF INSPECTOR with wide mechanical and electrical experience, preferably in tele-communications or light electronic engineering, required at once. Applicants should not be over 45 years and will be required to produce satisfactory evidence of their organizing, tech-nical and administrative ability in a similar capacity in a large engineering works. A salary commensurate with the responsibilities of the position is offered. Applications, which will be treated as strictly confidential, should be sent with full details of past career to Personnel Manager, The Phoenix Telephone and Electric Works Ltd., The Hyde, N.W.9. W 2494 CHIEF INSPECTOR required by well-known

Works Ltd., The Hyde, N.W.9. W 2494 CHIEF INSPECTOR required by well-known Company in S.E. London engaged in manufac-ture of light electrical appliances. Applicant must be technically qualified in electrical and mechanical engineering, and must have experi-ence of modern inspection methods for quantity production, including quality control and administration. Executive post carrying appro-priate salary and pension. Applications will be treated in strict confidence, but will not be considered unless they present a concise table showing education, training and positions held, with names, dates and salaries. Box No. W 2426.

W 2426. DECCA RADAR LIMITED require draughts-men and iunior draughtsmen for Research Drawing Office, preferably experienced in any of the following fields: Radar, radio and elec-tronic circuits, electro-mechanical devices, light mechanical engineering. Knowledge of work-shop practice essential. applicants must possess Ordinary National Certificate or equivalent. Positions permanent and progressive: salaries based on A.E.S.D. rates. Write giving fulf details to Chief Draughtsman. Decca Radar Ltd., 2, Tolworth Rise, Surbiton, Surrey. W 161

W 161 DECCA RADAR LTD. invites experienced Electrical Engineers to apply for the post of leader of a group, to be formed, for the pre-production of microwave links. The post carries good starting salary, is pensionable, and there are good prospects of expansion in scope. Appli-cants must be of British nationality and should be under forty years of age. Write, quoting ref. SRT.B.: The Research Laboratorics, 2, Tolworthe Rise, Surbiton, Surrey. W 168 DECCA RADAP LIMITED invites cardionicity

DECCA RADAR LIMITED invites applications DECCA RADAR LIMITED invites applications from experienced microwave engineers to join the Company in its extensive work in a wide field of microwave link and radar development. The Company offers excellent starting salaries and first rate opportunities for men to exploit their initiative and to rise rapidly to responsible posts. Graduates without industrial experience who are prepared to undertake intensive train-ing are also invited to apply for junior posts. Applicants must be of British nationality. Apply in writing, quoting ref. MW, to Research Director, Radar Laboratory, 2, Tolworth Rise, Surbiton, Surrey. DECCA RADAR LIMITED requires a senior

Surbiton, Surrey. W 170 DECCA RADAR LIMITED requires a senior engineer to take charge of the testing of high-power centimetric radar equipment. Previous experience of high-power radar, including receivers and A.F.C. systems, is required. A good salary, commensurate with experience, will be offered to the right man, for whom there are excellent prospects in this progressive and expanding company. Please apply giving full details of experience and stating salary required to Ref. F.L., Decca Radar Limited, 1-3, Brixton Road, London, S.W.9. W 2471 DECCA RADAR LIMITED require a Senior

Road, London, S.W.9. W 247t DECCA RADAR LIMITED require a Senior Engineer to form a new Division of the Com-pany undertaking the planning and erection of large radar stations both in the United King-dom and abroad. The appointment calls for a man between 30 and 45 years of age, with exceptional experience in the engineering design, planning, and progressing of electrical projects of a major nature. While a knowledge of radio and radar engineering at high powers is a dis-tinct advantage, very favourable consideration will be given to heavy power engineers. Sound engineering knowledge, experience, and ability

CLASSIFIED ANNOUNCEMENTS continued on page 4

The public acclaim these two units

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These units are deservedly popular — they meet a widespread demand for the highest quality at reasonable cost. They can be heard, together with the complete range of Speakers, at our London showrooms (109 Kingsway) from 9 a.m. to noon any Saturday, by appointment. Please write, or telephone Holborn 3074.

MODEL 1012 (10"). Generally regarded as the most outstanding speaker of its kind. Response: 55-11,000 c.p.s. Magnet of 12,000 gauss. Handling capacity 10 watts u.d.o. Price (tax paid) minus transformer £3-9-7

MODEL S.12135 (12"). For large handling capacity with balanced overall response, this unit is quite exceptional. Response: 50 - 11,000 c.p.s. Magnet of 14,000 gauss. Handling capacity 15 watts u.d.o. Price (tax paid) minus transformer £12 - 6 - 6

WHITELEY ELECTRICAL RADIO CO. LTD . MANSFIELD . NOTTS

## VITREOUS ENAMELLED WIRE WOUND RESISTORS

The following sizes are at present being manufactured :

Туре	Commercial Rating (watts)	Equivalent Services Rating (watts)	Nominal Diameter	Nominal Body length
LWV4-I	3	-	6 m.m.	16 m.m.
LWV4-J	4	. 3	6 m.m.	21 m.m.
LWV4-K	6	4.5	6 m.m.	34 m.m.
LWV4-L	10	6	6 m.m.	45 m.m.

ACCURATE • STABLE • RELIABLE • ROBUST

Please write for our illustrated leaflet No. E.E./LWV.

Labgear (Cambridge) Ltd.

WILLOW PLACE, CAMBRIDGE

PHONE 2494-5

#### SITUATIONS VACANT (Cont'd.)

The engagement of persons answering these advertisements must be made through a Local Office of the Ministry of Labour or a Scheduled Employment Agency if the applicant is a man aged 18-64 inclusive or a woman aged 18-59 inclusive unless he or she, or the employment, is excepted from the provisions of the Notification of Vacancies Order, 1952.

to plan and direct are the major requirements. A salary commensurate with the qualifications and responsibility will be paid. The post is permanent and pensionable. Apply in strict confidence to Managing Director, Decca Radar Limited, 1-3, Brixton Road, S.W.9. Ref. H.I.D. W 2470

W 24/0 DESIGNER / DRAUGHTSMAN required by The English Electric Co. Ltd., Luton. Must be fully experienced in radar or electronic equipment and conversant with RCS 1000 Standards. Good mechanical and circuit knowledge is essential. This is a responsible and progressive position leading a small team of draughtsmen. Salary around £700 with housing assistance for suitable married candidate. Applications to Dept. C.P.S., 336, Strand, W.C.2, quoting ref. 1119. W 2469

W 2459 DESIGNER DRAUGHTSMAN, Higher National Certificate Standard, preferably with machine tool experience, a knowledge of hydraulics and up-to-date machine tool practice, required by well-known Engineering Company. North-West Birmingham. Five-day week, Staff Pension Scheme, Modern Drawing Office. Apply, giving details of age, experience and salary required, to Personnel Manager, Box No. W 2485.

DESIGNER DRAUGHTSMAN. Electronic Instruments or Radio. Good prospects and salary according to qualifications. Prospects of housing assistance to suitable applicant. Saturday interviews if required. Near City centre and all amenities. Within easy reach London. Apply Marconi Instruments Ltd., Longacres, Hatifeld Road, St. Albans. W 2432

Road, St. Aioans. W 2432 DEVELOPMENT ENGINEER required. Experience in electronic communication field an advantage. Excellent opportuaity for man with technical and industrial experience to broaden experience with well-established and well-known Company. Near town centre and all amenities and within easy reach London. Apply Marconi Instruments Ltd., Longacres, Hatfield Road, St. Albans. W 2433

Aloans. w 2433 DEVELOPMENT ENGINEER required by well established firm to develop components and electronic apparatus. Ability to exploit ideas and work on own initiative essential. Send full details of previous positions held and type of work undertaken, together with approximate salary required, to Box No. W 2467.

DRAUGHTSMAN wanted for a small Drawing Office engaged in the design and drafting of Electronuc Instruments for the industrial and telecommunications field. Write stating age, experience and salary required to Dawe Instruments Ltd., Harlequin Avenue, Great West Road, Brentford, Middlesex. W 2482

Road, Brentford, Middlesex. W 2482 DRAUGHTSMEN, Senior and Junior, required. Design and Layout experience of Electronic and Electro-mechanical equipment and/or mechanical computors essential. Facilities exist for further technical training under the supervision of E.M.I. Institutes Ltd. Apply by letter stating age, nationality and experience to E.M.I. Engineering Development Limited, Wells Division, Penleigh Works, Wells, Somerset. W 2462

E. K. COLE LTD., Electronics Division, Malmesbury, Wiltshire, require Draughtsmen (Mechanical or Electrical of National Certificate Standard), aged 21 to 25 years, for interesting and varied work on Radar and Communication Equipments: also for Test Gear. Excellent conditions and prospects. Apply in writing to Personnel Manager. W 2458

ELECTRONICS. A vacancy exists with an Electrical Engineering Company on the North East Coast for an engineer to design and develop Electronic Circuits. Previous experience on this class of work essential. Applicants should submit full particulars regarding age, quslifications and practical experience to Box W 2435. ELECTRONIC CIRCUIT AND DEVELOP-MENT ENGINEERS are required by the English Electric Company Limited for interesting work on H.F. heating generators. A University Degree is desirable but practical experience of this type of work or on radio transmitters or similar apparatus is essential. Write with full details quoting .reference 357F to Central Personnel

Services, English Electric Co. Ltd., 336/7, Strand, London, W.C.2. W 2423 ELECTRICAL ENGINEER OR PHYSICIST, preferably with experience in high frequency in West Midlands, dealing with insulation and dielectrics. Candidates should have a University Degree or equivalent and be between 24 and 30 years of age. The post is a permanent and pensionable one and carries a maximum salary of £1,000 in the case of a well qualified and experienced applicant. Write to Assistant Staff Manager, Box No. W 2445.

Manager, Box No. W 2443. **ELECTRONIC ENGINEER** with H igher National Certificate or equivalent is required for the development of V.H.F. equipment and components. Applicants must have good laboratory experience and sound practical knowledge in this sphere. Write stating qualifications, experience and salary required to Wingrove & Rogers Ltd., Domville Road, Liverpool, 13. W 1720

ELECTRONIC ENGINEER required for development work in a laboratory (London S.E. Arca) engaged in the application of electronic techniques in new fields. Pension Scheme. Give full details of qualifications, experience, age and commencing salary required. Box No. W 2427. ELECTRONIC ENGINEER for work in connexion with the development and production of electronic measuring instruments. Sound theoretical knowledge and several years' experience in this field essential. Excellent opportunity for capable person to progress with young and rapidly expanding organization. Rivlin Instruments Ltd., 7a Maitland Park Villas, N.W.3. W 1724

W 1724 **ELECTRONIC ENGINEER** required by new division of prominent Engineering Concern in Northern Ireland for development work on Guided Weapons and other projects. Degree or equivalent in electrical engineering or physics, with good practical experience, preferably of D.C. Amplifiers, Electronic Computation, Pulse Techniques, or Miniature Equipment. Good salary and prospects for man with originality. Details of housing accommodation available and facilities for removal supplied at interview. Send full details of age, qualifications and experience, quoting reference E.E.1, to Box No. W 2487.

W 2457. **ELECTRONIC ENGINEER** required to work on development projects in small organization situated on South Coast. Experience on design of communications receivers essential. Some F.M. experience an advantage. Degree or equivalent desirable. Write full details and salary expected. Box No. W 1704.

salary expected. Box No. W 1704. **ELECTRONIC ENGINEER**, preferably graduate of recognized training establishment, required for design and maintenance of Industrial Electronic Control and testing equipment of all types. Salary according to qualifications and experience. Location, Midlands Engineering Plant. Apply Box No. W 1705.

Plant. Apply Box No. W 1705. **ELECTRONIC ENGINEER.** First-class man with commercial experience required by large organization. Applicant must have good basic training and preferably hold science Degree or membership of approved institute, and have sound knowledge and experience of electronics industry. Position offers excellent prospects for engineer aged 35/40 with initiative. Pensionable appointment. Minimum salary £1,000. Write giving full details of education, training and experience to 522, Box No. W 2490.

experience to 522, Box No. W 2490. **ELECTRONIC ENGINEER** required by large food factory in London, in the development section on the construction of specialized factory electronic equipment. Age 20-28. Applicants should possess initiative and be capable of working with the minimum of supervision. Good canteen. Salary according to age and experience. Write Box E.344, c/o 191 Gresham House, E.C.2. W 2500 ELECTRONICC EDICIDEER

Gresham House, E.C.2. W 2500 **ELECTRONICS ENGINEER.** Davy and United Engineering Company Limited, Sheffield, Europe's largest builders of rolling mills and forging presses, require the services of a young Physicist or Electronics Engineer for the development of electronic circuits for servo mechanisms and instruments for the Steel industry. A knowledge of industrial electronic instruments is essential. Salary will be commensurate with experience and qualifications. Staff Pension Scheme, good conditions of service and prospects. Apply, stating full particulars of experience and qualifications to the Personnel Officer. Davy and United Engineering Company Limited, Park Iron Works, Sheffield. 4. W 2499

ELECTRONIC ENGINEER/PHYSICIST required, primarily for component analysis. Appli-

cants should have experience of electronic measurements and instruments, and theoretical knowledge up to at least Inter. B.Sc. standard. Some knowledge of instrument design would be an advantage. The position offers a good salary and prospects of promotion to assistant chief engineer to a keen and conscientious man. There is also a vacancy for junior work on similar lines. Five-day week, W. London area. Our staff have been informed of this advertisement. Apply giving full particulars in confidence to Box No. W 2507.

ELECTRONIC INSTRUMENTS. Salesman required for South Africa; should have a sound knowledge of basic physical principles, preferably with University Degree and an interest in nucleonics. The successful applicant will receive training at the A.E.R.E. Harwell. Initial contract three years. Suitable salary and travelling expenses. Box No. W 1718.

ELECTRONIC INSTRUMENTS LTD., of Richmond, Surrey, have a number of vacancies for drawing office staff and for junior technical development engineers. Age: 21 upwards. Qualifications: National or Higher National Certificate in appropriate subjects. Excellent prospects in expanding and progressive company. Apply in person or by letter, giving experience. W 1707 ELECTRONICS RESEARCH ASSISTANT wanted in physiological laboratory. Duties: to develop and construct apparatus (mainly electronic) for nerve and muscle research. EEG/ EMG/CRO-photography or general workshop experience an advantage. Salary £520 × £25 to £620 p.a. with superannuation and children's allowance. Apply in writing to The Secretary, Department of Physiology, The Middlesex Hospital Medical School, London, W.1. W 1725

W 1/23 ENGINEER required to undertake the development of electronic instruments. The successful applicant must be capable of working substantially on his own imitative. Salary will be in accordance with qualifications and experience. Degree or equivalent preferred. Apply in writing to Advance Components Limited, Back Road, Shernhali Street, London, E.17. W 2466

Shernhali Street, London, E.17. W 2000 ENGINEERS: Mechanical, Electro-Mechanical and Electronic, required at Brentford and Feitham, Middlesex, for interesting development work on aeronautical and marine products and guided missiles. Applicants must have good technical qualifications together with appropriate practical experience. Pension scheme. Opportunities of advancement. Apply with full details of experience and salary required to Personnel Manager. Sperry Gyroscope Co. Ltd., Great West Road, Brentford, Middlesex, auoting No. 1098. W 2441

ENGINEERS, B.Sc. or equivalent, for design work in connexion with television circuits and associated components. Mechanical Engineers to undertake design of vibrators and/or switch and panel accessories. Good Degree or a wide industrial experience of this class of work. Engineers fully conversant with the design of transformers associated with Radio, Electronics and Telecommunications equipment. A large engineering Company invites applications for the above permanent and progressive appointments necessitated by expansion of the Company's business. Superannuation scheme in operation. Please write, in confidence, quoting reference EE/56, giving full details of salary , required to Box No. W 2491.

ENGINEER-DRAUGHTSMAN required for works West London Area. Good electrical qualifications and at least ten years' drawing office experience, including design of precision light mechanical equipment. Competent to initiate design and take charge of small drawing office engaged on interesting electro/ mechanical and electronic equipment, part of expanding company's engineering division. Pension scheme. Write stating qualifications, age, details experience, salary required. Box No. W 2452.

EXPERIENCED Fault-Finders wanted by Midland Manufacturers of Radio Equipment.

CLASSIFIED ANNOUNCEMENTS continued on page 6

## there are times when

. . . an airman feels on top of the world. The carpet of cloud makes him feel like a god of myth, high-stepping with power at his finger's control.

He may climb with effortless ease; he may dive like a dolphin at play; and he may pause to review the qualities of those who placed him on high.

Aerodynamicists, designers, craftsmen, and crew—confidence in these gives the pilot confidence in himself, in his instruments.

Serving those instruments in their turn are Parmeko transformers, earningby their steady, infinite reliability-the trust of those who plan.

Parmeko are proud of their part in the chain of confidence that keeps them flying.

## **PARMEKO** of LEICESTER

Makers of Transformers for the Electronic and Electrical Industries.



#### SITUATIONS VACANT (Cont'd.)

The engagement of persons answering these advertisements must be made through a Local Office of the Ministry of Labour or a Scheduled Employment Agency if the applicant is a man aged 18-64 inclusive or a woman aged 18-59 inclusive unless he or she, or the employment, is excepted from the provisions of the Notification of Vacancies Order, 1952.

Permanent posts located in the Midlands are offered to men with experience of Radar, Radio Control, V.H.F. Equipment. Write stating fully, experience and salary required to Personnel Manager. Box No. W 2318.

sonnel Manager. Box No. W 2010. EXPERIENCED Radio Testers and Inspectors required for production of communication and radio apparatus. A ls o Instrument makers, wirers and assemblers for Factory Test apparatus. Apply Personnel Manager, E. K. Cole Ltd., Ekco Works, Malmesbury, Wilts. W 146

W 146 FERGUSON RADIO CORPORATION LTD. have vacancies for the following staff. Development Engineers, Senior and Junior, for Radio, Television and Electronics. Previous experience circuit development work absolutely essential. Draughtsmen, experience in Radio and Television Receiver Design (and small mechanical details), able to prepare drawings for manufacture of prototype equipment. Apply in writing, Employment Manager, F.R.C., Gt. Cambridge Road, Enfield, Middlesex, Posts are permanent, progressive and pensionable. W 2480 GLASCOW, Padio Mechanic and features and the sentence of the sentence

GLASGOW. Radio Mechanic, preferably experienced Tape and Wire Recorders. Permanent position with prospects. Commencing salary 27 10s. 0d. Land, Speight & Co., Ltd., 73 Robertson Street, Glasgow. W 1726 GRADUATE in Electrical Engineering or Physics required for very interesting development of new materials, for use in Telecommunication. Apply Personnel Manager, Standard Telephones & Cables Ltd., North Woolwich, E.16, stating age, qualifications and salary required. W 2440

required. W 2440 GUIDED WEAPONS. Electrical Engineer or Mathematician with relevant postgraduate experience, required for theoretical work on design of automatic control systems. Some knowledge of effect of noise on servomechanisms an advantage. Senior staff appointment with good salary and prospects. Pension scheme. Write in detail to Personnel Manager, de Havilland Propellers Ltd., Hatfield, Herts. W 2480

W 2488 **IMMEDIATE APPLICATIONS** are invited for post as deputy head of electrical/radio office of leading Aircraft Company in the South. Applicant must have sound design experience including aircraft, with specialist knowledge of installation of electrical, radio, radar and electronic instruments and the development of ancillary apparatus. Applicant should be good disciplinarian. The post carries senior grading and housing can be recommended after a short probationary period. Please write in first instance giving experience, and salary expected, to Box N.2990, A.K. Advg., 212a, Shaftesbury Avenue, London, W.C.2. W 2431 HE INDUCTION HEATING CONTRACT

Avenue, London, W.C.2. W 2431 **H.F. INDUCTION HEATING CONTRACT** ENGINEER. Duties involve receiving customers' enquiries and/or samples, returning processed samples with report and equipment quotation. Acknowledging customers' order and issuing necessary information to works, D.O. and test room. Maintaining contact with customer' Ability to get on with other people and previous induction heating experience essential. Write giving age, past experience and salary required. Post in modern greenbelt factory with makers of one-third of all British R.F. heating plant. Radio Heaters Ltd., Wokingham, Berks. W 1715

W 1/15 INSPECTION SUPERVISORS required by a farge engineering Company situated in the London Area. Applicants should have held a similar position and have previous experience of radar and/or radio equipment. A knowledge of pulse circuity essential. Modern factory and good staff conditions. Please write, in confidence, quoting reference EE/IP, giving full details of experience to Box No. W 2443.

details of experience to Box No. W 2443. **LABORATORY ASSISTANTS** required on development of special materials for use in telecommunications. Applicants must have National Certificate or higher qualifications and a special interest in Chemistry or electrical measurements. Apply in writing to Personnel Manager, Standard Telephones & Cables, Ltd., N. Woolwich, E.16, stating age, qualifications :and salary required. W 2486 JOSEPH LUCAS (Gas Turbine Equipment) Ltd. invites applications for Electronic Engineers. Some knowledge of Servo mechanisms would be an advantage. Should have University Degree. These appointments are pensionable and offer good prospects to individuals with initiative and technical ability. Salary will be in accordance with experience and qualifications. Details of experience and qualifications should be sent to Personnel Manager, Joseph Lucas (Gas Turbine Equipment) Ltd., Shaftmoor Lane, Hall Green, Birtmingham. W 2459

LONDON BRICK COMPANY LIMITED has a vacancy for an Instrument Mechanic, preferably with some electronic experience, to work in the Research Department at Stewartby. Salary £400-£550 according to qualifications and experience. Apply in writing to the Personnel Manager, London Brick Company Limited, Stewartby, Bedford. W 1727

LONDON ELECTRICAL MNFG. CO. LTD. require Senior Production Engineers with theoretical and/or practical knowledge. Applicants should have had experience in the industry, preferably with components. Preference and considerably higher salaries would be offered to men who have been engaged in the production of ceramic condensers. Please write, giving full particulars, in confidence, to: L.E.M., Beavor Lane, W.6, or telephone Riverside 4824, asking for Mr. Lyons. W 2449

side 4824, assuing to accelerate the second second

McMICHAEL RADIO LTD. require Senior and Junior Engineers in their equipment division laboratory at Slough. Training and experience in the field of applied electronics (including communications) and experience of working with Government Departments are the chief qualifications required. Write stating age and full details of training, qualifications and experience to the Chief Engineer, Equipment Division, McMichael Radio Ltd., Slough. Bucks. W 157

MURPHY RADIO LTD. have vacancies for two or three Senior Engineers to lead development teams in their Electronics Division. Applications are invited from men with Engineering or Physics Degrees or equivalent, and with first class experience in the fields of Radar Navigational Aids, V.H.F. Communications Receivers and Low Power Transmitters. Salary upwards of £650 per annum according to qualifications and experience. Candidates prepared to bring energy and drive to their work may address their applications in confidence to the Personnel Manager, Murphy Radio Ltd., Wetwyn Garden City. W 2361

Garden City. W 2361 NEW DIVISION of the G.E.C. Stanmore Laboratories requires Electronic or Electrical Development Engineers for work on design of prototype models of equipment for guided missiles and radar. 1. Men with good experience of work in these fields are invited to apply for senior positions for which qualifications of degree standard or the equivalent would be an advantage. 2. A number of vacancies also exist for more junior men with qualifications to Ordinary National Certificate or equivalent standard. Applicants' requirements for holidays will receive special consideration. Applications should be made in writing to the Personnel Manager (Ref. EE/EI), stating age, qualificatories, Brown's Lane Division. The Grove. Stanmore Common, Stanmore, Middlesex. W 2493

PHYSICIST, first or second Honours Degree, required by electrical engineering Company in S.W. London. Preferably some knowledge of electronics. X-ray or nucleonics. Write stating age, experience and salary required to Box EE312, L.P.E., 110, St. Martin's Lane, W.C.2. W 2448 PHYSICIST required in the North West for the development of Capacitors for use in Telecommunication equipments, including silvered mica and paper and plastic dielectric types. The posts available are permanent, on full staff status with contributory Pension Fund, etc. Please write giving full details of qualifications and experience and the approximate salary required to Box 194, Dorland Advertising Ltd., 18/20, Regent Street, London, S.W.1. W 2489 PRODUCTION ELECTRICAL TESTERS required for television, radio and electronic equipment. Very interesting work. Good facilities and pay. Keen technicians only required. Dynatron Radio, Ray Lea Road, Maidenhead. W 2479

QUALIFIED MECHANICAL ENGINEER with electronic and light electrical experience required by Company engaged in design and development of guided weapons; preferably with knowledge of servo mechanisms to take charge of laboratory engaged on the design and development of control systems, power supplies, electrical gyros and other ancillary devices. Salary up to £1,500 p.a. for applicant with suitable qualifications, Pension scheme; housing assistance. Apply Box AC 80899, Samson Clarks, 57-61 Mortimer Street, London, W.1. W 2422

RADIO AND RADAR TESTERS. First-class men required for work on V.H.F. Communication Gear and Government Contracts for Radio and Radar Equipment by Midland Manufacturers. Men with wide experience of Fault Finding in any of the fields mentioned should write giving full details to Box No. W 2317.

**RECORDING STUDIOS** specializing in the production of commercial discs and radio programmes require the services of an experienced balance engineer who is fully conversant with disc and tape recording techniques. A good technical knowledge of the equipment used in this type of work is essential. Applications in writing giving age, experience and salary required to Box No. W 1713.

writing giving age, experience and salary required to Box No. W 1713. **RESEARCH AND DEVELOPMENT ENGIN-EERS** are required by British Telecommunications Research Ltd., a Company associated with The Automatic Telephone & Electric Co. Ltd., and British Insulated Callender's Cables Ltd., for work on long term development projects in Wide-band Line Communications. A number of posts with salaries in the range £500-£1000 per annum are available for suitably qualified engineers or physicists with experience in this field. Further posts are available for technical assistants with salary in the fange 5300-£600 according to qualifications and experience. Applications are also invited from Honours Graduates in physics or electrical engineering who are considering careers in the research or development side of the telecommunications industry. There is a superannuation scheme and the Company works a five-day week. Application should be made to the Director of Research, British Telecommunications, gesearch Ltd., Taplow Court, Taplow, Bucks, giving age and full details of education, qualifications. experience and approximate salary required. **W2463 RESEARCH ENGINEERS** required for automatic welding development projects. Applicants

RESEARCH ENGINEERS required for automatic welding development projects. Applicants should possess the necessary aptitude and qualifications to undertake research work—a Degree in electrical engineering is essential—but ample opportunity is available for outstanding successful candidates to make further progress in the automatic welding field, in which this North-East firm occupies an important position. Apply in writing to Box No. W 1688.

SALES ENGINEER required for the sale in Great Britain of world known test gear in the electronic field. Write stating experience and suggestions to Box No. W 1706.

suggestions to Box No. W 1/06. SENIOR, Intermediate and Junior Draughtsmen required by The English Electric Co., Ltd., Luton, for work on design of specialized electronic equipment. Applicants should have sound mechanical background and an interest in electrical or electronic work. Part time study for National Certificate encouraged. Applications from practical engineers without Drawing Office experience will also be considered for full time training course. Remuneration according to

CLASSIFIED ANNOUNCEMENTS continued on page 8



## **O-1000 VOLTS** with the new **SOLARTRON** Twin-Regulated Power Supply

MODEL SRS. 152

### FEATURES:

Wide range of positive and negative output voltages. High quality components including only paper type capacitors. Automatic overload protection. High stability.

## SPECIFICATION:

D.C. Output Series Condition : 0-1,000 volts +ve or -ve, 0-150 m/a. or 0-500 volts +ve and 0-500 volts -ve, 0-150 m/a.

Parallel Conditions : 0-500 volts +ve or -ve, 0-300 m/a. A.C. Output Unstabilised :

Two outputs 6.3 volts, 5 amps. each. Stability :

0.2% for  $\pm 10\%$  mains input change.



The units are available separately for bench or 19 in. rack mounting.

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MODEL SRS.151. Regulated Power Supply. Output 20-500 volts at 0-300 m/a. Separate negative output 0-170 volts. Two 6.3 volt A.C. outputs at 5 amps each. MODEL SRS.153. Solartron Vari-Pack. Output 0-500 volts positive or negative unregulated at 100 m/a. 4 or 6,3 volts A.C. output at 3 amps.

STAND No. 10G

BRITISH INSTRUMENT INDUSTRIES EXHIBITION JUNE 30th to JULY 11th

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Write for details to :

SOLARTRON LABORATORY INSTRUMENTS LTD.

22 HIGH STREET KINGSTON SURREY. TEL.: KINGSTON 8981 P.B.X

#### SITUATIONS VACANT (Cont'd.)

The engagement of persons answering these advertisements must be made through a Local Office of the Ministry of Labour or a Scheduled Employment Agency if the applicant is a man aged 18-64 inclusive or a woman aged 18-59 inclusive unless he or she, or the employment, is excepted from the provisions of the Notification of Vacancies Order, 1952.

qualifications and experience. Applications to Central Personnel Services, The English Electric Co. Ltd., Marconi House, 336/7, Strand, London, W.C.2, quoting reference 144J. W 2439

W 2439 SENIOR Mechanical designer. Two or three vacancies exist in our Electronics Division for Senior Design Draughtsman on Electronic Equipment. Candidates should be of at least Higher National Certificate standard and should be capable of leading a design team in this class of work. Salary £650 per annum upwards according to qualifications and experience. Applications should include full details of experience to date and may be forwarded in confidence to the Personnel Manager, Murphy Radio Ltd., Welwyn Garden City. W 2360 SENIOR Test/Service Engineer required for

Radio Ltd., Welwyn Garden City. W 2360 SENIOR Test/Service Engineers required for Industrial Electronics Department. H.N.C. an advantage. Experience in electronic control gear, H.F. heating equipment or instruments preferred. Please apply, stating age, qualifications and experience, to Central Personnel Services, English Electric Co., Ltd., 336/7. Strand, London, W.C.2, quoting reference 979B. W 2424

SCIENTIFIC INSTRUMENT MAKERS in North London have a few vacancies for skilled laboratory assistants and wire-men for the development and production of high quality electrical and electronic apparatus. Please apply. in writing only, stating qualifications. experience and salary required, to Personnel Officer, Hilger and Watts Ltd., Hilger Division, 98 St. Pancras Way, London, N.W.1. W2410

SENIOR DESIGN DRAUGHTSMEN required by a large engineering company due to expansion of company's services and commercial business. Applicants should have a comprehensive knowledge of mass production of radio, television and Services equipment. Salary £650 per annum: monthly status. The appointments are of a permanent and progressive nature. Company superannuation and life assurance scheme in operation. Please reply, quoting reference EE/53, giving details of experience to Box No, W 2460.

Sention No. W 2460. SENIOR RESEARCH ENGINEER required to supervise gyroscopic development programme. Applicants must possess first rate academic qualifications and must have had considerable experience of laboratory research techniques. together with detailed experience of practical engineering design and industrial manufacturing processes and k n o w l e dg e of engineering materials. Understanding of instrument design techniques, servo mechanism theory, and electronics desirable. Company is situated West side of London. Apply with full details, including salary, quoting No. 1100, to Box AC81308, Samson Clarks, 57/61, Mortimer Street. W 1. W 2450

W 2450 SEVERAL EXPERIENCED graduate electronic and electro-mechanical engineers are required for long term development in centimetric radar equipment, servo mechanisms and instrumentation. Candidates must be British born and should be capable of working as members of a team with little supervision. The posts are permanent and pensionable. Applications, giving full particulars of experience to date, will be treated in confidence and should be made in writing, quoting reference SP, to the Personnel Manager, The Plessey Company Limited. Vicarage Lane, liford. Essex. W 2425 TELEPHONE ENGINEER required for name

Manager, The Piessey company Linucy, Vicarage Lane, Ilford, Essex. W 2425 TELEPHONE ENGINEER required for panel layout and component engineering work on carrier-current equipment. London S.E. area. Qualifications: up to Higher National Certificate, and experience of transmission of electronic apparatus work desirable. State age and commencing salary required. Box No. W 2475. TEST AND JUNIOR Development Engineers are required to assist in the development and production of precision electronic laboratory instruments. Applicants should have a sound theoretical knowledge and have had substantial practical experience of measuring instrument circuits, Salary in the region of £450-£650 p.a. according to qualifications and experience. Apply stating full details to the Chief Engineer, Furzehill Laboratories Ltd., Boreham Wood, Herts. W 167 SIR W. G. ARMSTRONG WHITWORTH AIRCRAFT LTD. Vacancies exist for three Electronic Technicians with Degrees or equivalent qualifications for work in connexion with Flight Trials of Guided Missies. Experience on the Service Radio and Radar equipment would be an advantage. Applications should be addressed to the Chief Engineer, Sir W. G. Armstrong Whitworth Aircraft Ltd., Baginton, nr. Coventry. W 2331

THE BRITISH IRON AND STEEL Research Association. Young Electronics Engineer required by the Plant Engineering Division of the above Association to assist in instrumentation work involving use of strain gauges and multichannel metering and recording circuits in connexion with new developments in steelworks plant. Applicants should have an Ordinary or Higher National Certificate in electrical engineering. Age range 20-24. Starting salary according to age, qualifications and experience within the range f430-f480 p.a. Location of work London, with visits to steelworks. Written applications only quoting "Instruments" to Personnel Officer, B.I.S.R.A., 11, Park Lane, London, W.I. W2444

THE DE HAVILLAND ENGINE CO. LTD. Electronic Engineer required, interested in physical measurement rather than construction of apparatus, required for vibration measurement and analysis on gas turbine and piston engines. Please write in confidence stating age and full details of previous experience to the Personnel Officer, The De Havilland Engine Co. Ltd., Stag Lane, Edgware, Middlesex. W 2386 THE ENGLISH ELECTRIC Co., Ltd., Luton, have vacancies for Electronic Engineers for development work on V.H.F. Radio Subminiature equipment and/or recording techniques. Some Field Trials engineers and assistants also required. Applications stating age, experience and qualifications, and quoting ref. 456L, should be sent to Central Personnel Services. English ELECTRIC Co., Ltd., Starnd. London, W.C.2.

London, W.C.2. W 2504 THE ENGLISH ELECTRIC CO. LTD., Luton, have vacancies for Electronic Engineers for work in Australia on V.H.F. Sub-miniature equipment with Trials applications. Assistance with housing will be provided. Applications stating age, experience and qualifications should be sent to Central Personnel Services, 336/7, Strand, London, W.C.2, quoting ref. 456. W 2468

THE ENGLISH ELECTRIC CO. LTD., Stafford, requires the following personnel for its new factory in the North Staffs area: 1. Engineer with experience of magnetic amplifiers; good academic and practical background is essential. 2. Engineer with experience on electronic servo control gear and servo analysis. Housing facilities may be available in 12 months. Salary up to £1,000 per annum. Apply to Department Central Personnel Services, 336, Strand, W.C.2, quoting ref. 1095. W 2457

THE GENERAL ÉLECTRIC CO. LTD., Brown's Lane, Coventry, requires Senior and Junior Electronic Development Engineers for work on Guided Weapons and like projects, particularly in the field of Microwave and Pulse Applications. Mechanical Development Enginers, Designer Draughtsmen and Draughtsmen, preferably with experience of Radar type equipments, also required for the above projects. Salary according to age, qualifications and experience. Houses available for selected senior to the Personnel Manager (Ref. R.G.). W 169

THE GENERAL ELECTRIC CO. LTD., Brown's Lane, Coventry, requires Senior and Junior Electronic Development Engineers for work on Guided Weapons and like projects, particularly in the field of Microwave and Pulse Applications. Mechanical Development Engineers, Designer Draughtsmen and Draughtsmen, preferably with experience of Radar type equipments, also required for the above projects. Salary according to age, qualifications and experience. Houses will be allocated to selected staff. Apply by letter stating age and experience to The Personnel Manager (Ref. R.G.). W 158 THE PLESSEY COMPANY have vacancies at their llford Works for Engineers and Draughtsmen of senior and intermediate grades for development work on components and mechanisms for radio, television and communications equipment. These vacancies occur as a result of the enlargement of the laboratories to permit a substantially increased interest in the design of circuits and components for specific applications. All positions are permanent and pensionable and the expansion now in progress offers good prospects of advancement. Attractive initial and progressive salaries are offered to men who are qualified by reason of educational attainment or practical experience. All applications will be dealt with in confidence and should be addressed, in the first instance, quoting reference EE/ED, for the attention of the Personnel Manager, The Plessey Company Limited, Vicarage Lane, Ilford, Essex. W 2481 TECHNICAL JOURNALIST. The General Electric Co., Ltd., has a vacancy for a technical editorial assistant in the Press Relations Section of its Publicity Organization to undertake work in connexion with the Company's activities in the fields of telecommunications and electronics. Candidates should be between the ages of 25 and 38 and should have a good knowledge of radio and television techniques. Experience in journalism and ability to prepare interesting well-written article essential. Application should be made in the first place to Personnel Dept., Magnet House., Kingsway, W.C.2, stating age, education, experience and salary required. Samples of work should be submitted. W 2446 **TECHNICAL WRITER** required for Publicity Department of leading radio valve and elecronic tube manufacturers. Advertising experience an advantage. Age 26-30. Apply stating alary required to 487 Box No. W 2455.

**TECHNICIAN**, with experience in electronic work, required for development laboratory in large Telecommunication Engineering Works. Give particulars of experience, education and technical training, qualifications and commencing salary required. London S.E. Area. Box No. W 2426.

W 2426. **TEST EQUIPMENT DESIGNER** required to take charge of department producing handoperated and automatic test equipment for capacitor production. Must have good technical qualifications. Previous experience of massproduction testing desirable. Reply giving full details of age, technical training, past experience and salary required to A. H. Hunt (Capacitors) Ltd., Bendon Valley, Garratt Lane, Wandsworth, London, S.W.18. W 2484

Wandsworth, London, S.W.18. W 2484 VACANCIES are available in the Transmission Department for: (1) Development engineers, experienced in the design of multi-channel carrier telephone systems, or in allied techniques including amplifiers, filters and radio. (2) Equipment engineers, with such knowledge of modern line communication practice as will enable them, after training in the company's methods, to prepare detailed specifications of equipment requirements and to develop arrangements for rack-mounting of the apparatus. (3) Draughtsmen to work with the engineers, preferably having experience in the telecommunications industry. The positions available are permanent, on full established staff status, with contributory Pension Fund and usual staff conditions. Please write to Personnel Manager, Automatic Telephone & Electric Co. Ltd., Liverpool 7, giving full details of experience, qualifications, age and approximate salary sought. W 2429

mate salary sought. W 2492 WALVE ENGINEERS required for an expanding Production Unit in Scotland which is making special valves and gas-filled tubes for Radar. Applicants must have design or production experience under one of the following headings: Thermionic Valves. Microwave Valves. Gas-filled Tubes. Radar Equipment. Microwave Measuring Equipment. The positions to be filled are: Senior Engineer. Salary £1,000-£1,500 per annum according to experience and ability. An Honours Degree in Physics or Engineering is essential, and applicants must have held previously a position of responsibility. Junior Engineers-Salary range £500-£1000 per annum according to experience and qualifications. These positions offer full scope for initiative and progress. Contributory Pension Scheme in operation. When applying, give full details of

CLASSIFIED ANNOUNCEMENTS continued on page 10

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## The natural

## radioelements



RADIUM \* RADON RADIUM D POLONIUM MESOTHORIUM RADIOTHORIUM THORIUM X THORIUM B

Even in these days of man-made isotopes the historical natural radioactive elements are still indispensable to the physicist, the radiotherapist and the technologist in industry. Here at the Radiochemical Centre we have a long experience of these materials and are now applying to their production many of the new techniques developed in connection with pile-produced isotopes. Thus we are able to supply the natural radioelements in a purer state and a wider range of physical forms than ever before.

## as α-emitters, as β-emitters as γ-emitters, as neutron sources, as emanating sources

 $\star$  (Appliances containing over 1 mg. of radium are supplied to United Kingdom institutions only on loan. For owners or suppliers of radium elsewbere we supply appliances and undertake their loading. We also recover radium from old appliances.)

For further information please write to :-

THE RADIOCHEMICAL CENTRE, AMERSHAM, BUCKINGHAMSHIRE, ENGLAND

TAS/RC.5

SITUATIONS VACANT (Cont'd.)

The engagement of persons answering these advertisements must be made through a Local Office of the Ministry of Labour or a Scheduled Employment Agency if the applicant is a man aged 18-64 inclusive or a woman aged 18-59 inclusive unless he or she, or the employment, is excepted from the provisions of the Notifica-tion of Vacancies Order, 1952.

qualifications and experience in chronological order, quoting reference VE, to Box No. 938 B McCallum Advertising, 21 Castle Street, Edin-burgh 2. W 2476

VICKERS - ARMSTRONGS LIMITED have vacancies in their Guided Weapons Department for Inspectors on inspection and test of complex electronic devices; thorough experience of electrical laboratory instruments and low power V.H.F. technique essential; knowledge of Servo Systems would be an advantage. Apply in writing to the Employment Manager, Vickers-Armstrongs Limited (Aircraft Section), Weybridge, Surrey. Applications, with certain exceptions, are subject to the approval of the Ministry of Labour and National Service. W 2418

VICKERS - ARMSTRONGS LIMITED have vacancies in their Guided Weapons Department for the following staff. (1) Electronic Engineers of degree standard, with at least three years' experience in one of the following: V.H.F. Transmitter and Receiver Design, Pulse Tech-niques, Aerials, Transformers Design, Servo-Mechanisms and Electro-mechanical Devices. (2) Technical Assistants with H.N.C. or C. & G. standard, for development work, electronics, servo-mechanisms and hydraulics. (3) Draughts-men with experience in electronic, electrical or electro-mechanical design. (4) Jig and Tool Draughtsmen. Apply-quoting reference G.W.1 —to Employment Manager, Vickers-Armstrongs Limited, Weybridge Works, Weybridge, Surrey. Applications, with certain exceptions, are sub-ject to the approval of the Ministry of Labour & National Service. W 2454 YOUNG ENGINEERS with some experience

YOUNG ENGINEERS with some experience YOUNG ENGINEERS with some experience in television design and development are required for this type of work by a well-known Company in the North West of England. Ability to initiate and carry out development projects and a University Degree or equivalent are the essential qualifications. Arrangements will be made to house the successful applicant. Apply quoting reference C.I.H. to Box No. W 2453.

A further classified advertisement appears in display style on page 78.

#### SITUATIONS WANTED

CHIEF ENGINEER, ten years with present company, seeks similar appointment with ex-panding organization. Outstanding experience of development and production of airborne radio, servo-mechanisms, test equipment, etc. In-terested in hard work and a high salary. Box No. W 1716.

ELECTRONIC ENGINEER, age 33, B.Sc., Grad.L.E.E., experience-41 years R.E.M.E., 4 years in industry, requires post as technical representative for North East Coast area. Box No. W 1709.

No. W 1709. **ELECTRONIC ENGINEER** with several years' experience in guided weapon field, including development, test and considerable trials experi-ence, seeks position in same field. Responsibility more important than salary, but position pre-ferred as sponsor or trials leader. Used to arduous hours, and enthuisam for problems involved. Willing to go to Australia. Academic education extends to full technological standard, City and Guilds, Radio and Telecommunication Engineering. Box No. W 1717.

KEEN ELECTRONIC ENGINEER seeks change In employment. Experienced Public Address, Staff Location, Remote Control and allied systems; some telephone development experi-ence, competent draughtsman. Excellent refer-ences. Box No. W 1723.

## FOR SALE

AMERICA'S famous magazine Audio Engineer-ing, 1 year subscription 28s. 6d.; specimen copies 3s. each. Send for our free booklet

quoting all others: Radio Electronics, Radio and Tele. News, etc. Willen Limited (Dept. 9), 101 Fleet Street, London, E.C.4. W 108 CAMBRIDGE MODEL B two-way Recorders, 1, 0 to 1700°C and 1, 0 to 1400°C, also 4 Fery Radiation Pyrometers with adjustable stands. Details on request. Mullard, Blackburn. Works, Blackburn. W 1703

**ELECTRONIC COMPONENT SUPPLIES.** We specialize in the supply of Electronic Compo-nents, Accessories, Test Equipment, etc., for Government Depts., Industrial Concerns, Research Establishments, Laboratories, Colleges, etc. Your enquiries and orders will receive our prompt attention. Holiday & Hemmerdinger Ltd., 74/78 Hardman Street, Deansgate, Manchester, 3. Tel.: Deansgate 4121. W 148

FOR SALE—surplus to requirements. Model No. 1035 Double Beam Oscil Box No. W 1692, Cossor Oscillograph

FREE. Brochure giving details of courses in Electrical Engineering and Electronics, covering A.M.Brit, I.R.E., City and Guilds, etc. Train with the Postal Training College operated by an Industrial Organisation. Moderate fees. E.M.I. Institutes, Postal Division, Dept. EE29, 43, Grove Park Road, London, W.4. (Associate of H.M.V.). W 2808

MAGSLIPS at 1/10th to 1/20th of list prices. Huge stocks. Please state requirements. K. Logan, Westalley, Hitchin 1744, Herts. W 116 ments. K. W 116 MINIATURE BALL BEARINGS, Steel Balls, Stainless Steel and Phosphor Bronze Balls, Prompt delivery. Distributors: Insley (London) Ltd., 21/22 Poland Street, London, W. 1 Tel.; GERrard 8104 and 2730. W 155

SINE-COSINE Potentiometers, Magslips, Selsyns, Ipots, Velodynes and equipment and components for automatic control. Servotronic Sales, Inc. Hopton Radio, No. 1 Hopton Parade, Streatham High Road, London, S.W.I6. STReatham 6165. See our advertisement on page 80. W 2502

SINE-COSINE RESOLVERS (3" Magslip Transmitters No. 5, AP 10861). Brand new, each in maker's tin. Offered in quantity at less than one tenth of cost. Export inquiries in-vited. P. B. Crawshay, 166 Pixmore Way, Letchworth, Herts. W 153 WEBB'S 1948 Radio Map of the World, new multi-colour printing with up-to-date call signs and freah information; on heavy art paper 4s. 6d. post 6d. On linen on rollers 11s. 6d., post 9d. W 102

#### **EDUCATIONAL**

CITY & GUILDS (Electrical, etc) on "No Pass-No Fee" terms. Over 95 per cent suc-cesses. For full details of modern courses in all branches of Electrical Technology send for our 144-page bandbook-Free and post free. B.I.E.T. (Dept. 337C), 17 Stratford Place, London, W.1. W 142

London, W.1. W 142 CITY OF COVENTRY Education Committee. Coventry Technical College. Session 1953-54. Electronic Engineering. Applications are invited for entry to the next 3-year full-time course commencing in September, 1953, from those requiring a comprehensive training to an advanced level in Electronic Engineering, to qualify them for technical posts in radio, tele-communications, television and industrial elec-tronics. The syllabus will cover the requirements of C. & G., Brit.I.R.E., and I.E.E. examina-tions. Entry age 16 years or over. Application forms and further information available from the Principal, Coventry Technical College. W. L. Chinn, M.A., Director of Education, Council House, Coventry. BOROUGH OF BOLTON-Educa-

House, Coventry. W 1708 COUNTY BOROUGH OF BOLTON-Educa-tion Committee. Bolton Technical College. A three-year full-time course in Electronic Engin-eering commences in September, 1953. Appli-cants should be in the age range 16 to 18, and have obtained, or be taking, General Certificate (Ordinary Level) in Mathematics or Physics, or equivalent courses in technical institutions. This rapidly developing industry offers new and attractive openings to qualified men. Applica-tion forms and particulars may be obtained from the Principal, Technical College, Bolton, Lancs. W. T. Selley, Chief Education Officer. Education Offices, Nelson Square, Bolton. W 1658

PATENT

THE PROPRIETORS of British patent No. 645,975 concerning an electronic instrument for determining the degree of compaction of con-crete mixes in the fluid state, desire to contact any firm interested in the purchase rights, or obtaining grant of licences hereunder. Purther particulars from Box No. W 1728.

#### SERVICE

CONSTRUCTION of experimental and Labora-tory equipments in Radio and Electronics can be undertaken by a small group of engineera. Box No. W 1721.

ENGINEER. 10 years' research experience, H.N.C., wishes to offer services as consultant to small manufacturer on a freelance basis. Experienced in V.H.F., audio tape recorders, etc. Own W/S and drafting effort available. Box No. W 1722.

MANUFACTURES. As specialists in all types of Sheet Metal Work for the Radio and Elec-trical Trade we are able to quote you very keen prices for a high quality job. Intensalite Sign and Fixture Co. Ltd., 89, Leopold Street, Bir-mingham, 12. Telephone Vic. 0065. W 1684

METALWORK. All types cabinets, chassis, racks, etc., to your own specifications. Philpott's Metal Works Ltd. (G4B1), Chapman Street, Loughborough. W 2197

REPAIRS AND MAINTENANCE to all types of electronic equipment and precision instru-ments, Enquiries invited from industrial users, government depts., research and educational establishments, etc. Glyndon Radio Service Ltd., Industrial Equipment Department, 68, Wellington Street, S.E.18. Phone: WOO 2749. W 1659



In this new price list are given details of a carefully balanced range of transformers and chokes covering some

#### 100 STANDARD TYPES

There is now a Partridge Transformer-(Power or Audio) and Choke to fulfil virtually every need, of replacement or original circuitry. Each type is available for

IMMEDIATE DELIVERY and each is covered by a

12 MONTHS GUARANTEE

Logically the outstanding performance of Partridge products implies that they cannot be the cheapest available, but there is that assurance of long-life efficient service for so very little extra. Write for today's most comprehensive list of tried and proven transformers and chokes.





## ELECTRONIC EQUIPMENT

BTH pioneered the development and application of electronics in industry and now make a wide range of equipments including:—

Emotrol Variable-speed Motor Drives up to 750 H.P. Automatic Voltage Regulators Marine and other types of Radar High-frequency Heaters of the dielectric and induction types Resistance-welding Controls and Timers Photo-electric apparatus for detection, counting, register control, etc. Servo-controls Specialised Test Equipments





Member of the AEI group of companies



The valves — special and industrial types —needed in these equipments are designed and manufactured entirely by BTH and include :—

Magnetrons TR and TB cells Thyratrons — mercury and gas-filled types Hot-cathode Rectifiers Photo-conductive Cells Ignitrons Spark Gaps Germanium and Silicon Diodes & Rectifiers

BTH also make a range of apparatus for more specialised applications such as :--

Magnetic Sheet Testers

Gaussmeters

Mass Spectrometer & Halogen Leak Detectors

Over half-a-century's experience in the manufacture of all kinds of electrical products ensures that BTH electronic equipments are correctly designed and built to suit any industrial applications.

THE BRITISH THOMSON-HOUSTON COMPANY LIMITED, RUGBY, ENGLAND

## A4531

**JUNE 1953** 

## **POTTED IN** 'ARALDITE' – PROOF AGAINST CORROSION





'ARALDITE' is particularly valuable for potting transformers, capacitors, motor windings and a wide variety of radio and electronic apparatus. It has made "miniaturisation" more practicable; enclosing the equipment in a protective cover it provides complete insulation and offers good resistance to "tracking". High temperatures, humidity and other exacting requirements are not inimical to its use.



'ARALDITE' is a hot-setting ethoxyline resin proved by experience to be eminently suitable for large-scale production. Its shrinkage during casting is very low, as no water or volatile products are given off during setting.

## MADE IN OUR DUXFORD FACTORY-**DELIVERY FROM STOCK**

adhesives (Araldite Types 1

and 15) for bonding metals,

Summer and the second s

ARALDITE, is also available in

hot - setting

We shall be glad to send full details of properties, casting temperatures and curing times.

## **'ARALDITE' Casting Resin B**

porcelain, quarte, etc., and AS A SURface coating reain (Araldite 985 E) for use as a primer or lacquer for light alloys. -a product of AERO RESEARCH LIMITED A CIBA COMPANY DUXFORD · CAMBRIDGE 'PHONE: SAWSTON 187

**ELECTRONIC ENGINEERING** 

**JUNE 1953** 

# **\*ENGLISH ELECTRIC'** 'C' type Gores



## for use in Transformers and Chokes to RCS. 214

Details of these 'C' type Cores will be supplied on request to Transformer Sales and Contracts Department, East Lancashire Road, Liverpool 10.

THE ENGLISH ELECTRIC COMPANY LIMITED, QUEENS HOUSE, KINGSWAY, LONDON, W.C.2. WORKS: STAFFORD · PRESTON · RUGBY · BRADFORD · LIVERPOOL · ACCRINGTON

**JUNE 1953** 

Waveguide Impedance Measurement

## Unique features of design give extreme accuracy

Metrovick Standing Wave Detector

This instrument consists of a waveguide section built up from solid material, and a moving carriage, on which is mounted a tunable crystal detector.

The carriage is of the non-contacting type, and is located directly from the inside surface of the waveguide, thus eliminating any variation in probe penetration along the travel of the carriage.

This unique feature makes possible extreme accuracy in the measurement of waveguide impedances.

It has been designed for operation over the band of wavelengths from 10 cms. to 11 cms. in size 10 standard waveguide and it is fitted with Inter-Service Standard waveguide connectors.



METROPOLITAN-VICKERS ELECTRICAL CO. LTD., TRAFFORD PARK, MANCHESTER 17 Member of the A.E.I. group of companies

METROVICK Test gear for the microwave laboratory

R/E 204



## The Saunders-Roe Displacement Indicator

This instrument and its associated pick-offs have been designed to afford a degree of stability and linearity of calibration at least as good as that of a mechanical extensometer. In addition the maximum

possible measuring flexibility and adaptability have been incorporated. It is normally supplied for two channel working but additional channels up to a total of six can be made available if desired. Displacement is indicated as deflection on a panel mounted 0-1 milliameter and additional facilities are provided to connect the output stages to suitable direct recording equipments.

## THE PICK-OFFS

In order to secure linearity of calibration within a wide working range, the pick-off unit consists electrically of a constant impedance inductive half-bridge unbalanced by movement of a magnetic material in proximity to the inductive elements. Whilst this system is considerably less sensitive in terms of mechanical displacement/electrical output than the conventional differential transformer, it does provide extremely good linearity over a wide range, which is considered to be a prime feature in this type of equipment. The pick-offs may be used singly or in pairs. Paired use is a particularly valuable application, for instance, in process control, where dimensional monitoring is required in the absence of a convenient datum, e.g. plastic extrusion, wire drawing, foil rolling, etc.

An Inductive Pick-off

## **GENERAL SPECIFICATION**

The instrument is engineered to the standard of robustness necessary for general industrial use. All electronic circuits are carefully stabilised against mains supply variations, and components are of high quality and stability. ALL SALES ARE SUBJECT TO THE COMPANY'S STANDARD CONDITIONS.

#### INSTRUMENT

 Dimensions: Height: 10.5 in. Length: 17 in. Depth: 9 in.

 Weight: ... ... ... ... ... ... ... 33.5 lb. (15.3 kg.)

 Power Supply: ... ... 115-230 volts, 50-60 c.p.s. A.C.

 Measuring frequency and range: ... Max. 600 c.p.s.

 Measuring ranges:... ... 0-.005", 0-.015", 0-.050".

 Other ranges available: ... ... ... ... ... 0-0.10"

Dimensions: Mounting flange:	 1.30	Ler in.	ngth: dia.	2.3	5 in.	Dia: .90 in.
Weight:						57.5 grams.
Weight of moving elen	ient			•••	•••	2.5 grams.

PICK-OFF

## SAUNDERS BOE LTD ELECTRONICS DIVISION

OSBORNE . EAST COWES . ISLE OF WIGHT

# The "Belling-Lee" page for Engineers





L. 732 TWIN NEON CARRIER WITH RETAINING CLIP

L. 731 AS L. 730 BUT WITH BACK CONNECTIONS

L. 730 TWIN FUSEBOX WITH CARRIER AND RETAINING CLIP

## TWIN NEON INDICATING FUSEHOLDER

The twin neon indicating fuseholder L732 illustrated above top right, carries two neon lamps, and where required, may be used to replace the normal fuse carrier on standard "Belling-Lee" twin fuseboxes L1033 shown below. The overall depth is thereby increased by  $\frac{1}{2}$  in., but the panel area remains unchanged.

Two lamp-holders with resistors are built into the moulding, and are arranged in such a manner that the adjacent lamp gives visual indication when either fuse



## L. 1033/C3 WITH LID RETAINING CLIP

L. 1033/C4 AS ABOVE, WITH BACK CONNECTIONS & BUSHES blows. The neons will glow satisfactorily over a voltage range of 180/250v. a.c. (r.m.s.), or 220/250v. d.c.

A replacement retaining clip is supplied with the carrier. This ensures that when the box is properly wired, the neons can only be inserted in the correct phase of the supply.

Supplied separately or complete with bases.



CAMBRIDGE ARTERIAL RD., ENFIELD, MIDDX., ENGLAND

JUNE 1953

a publication of interest to scientific and industrial electronic engineers

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## present THEIR LATEST DEVELOPMENTS CONDENSERS...

FOR the benefit of those who were unable to visit the recent R.E.C.M.F. Exhibition, and for those who had not the time available to inspect our Stand in detail, we illustrate here the range of T.C.C. latest developments displayed thereon.

These are indicative of the progress when is constantly being maintained, and "which ensures that "T.C.C. Leadership in Condensers" is a tangible reality.

## I. SMALL CAPACITY CLOSE TOLERANCE TUBULAR CERAMICS

For top end coupling in Band Pass Filters. Capacity range 0.5 pF to 5 pF at 500 v. D.C. Tolerances  $\pm 20\%$  and  $\pm 10\%$ .

#### 2. TANTALUM ELECTROLYTICS

The special neutral electrolyte prevents corrosive injury in the event of mechanical damage. These 8  $\mu$ F condensers are for working voltages up to 120 v D.C. and for temperatures up to 120°C.

## 3. "METALPACK" & "METALMITE" PAPER TUBULARS IMPREGNATED in "VISCONOL-X"

New impregnant improves reliability at 100°C. Full Ministry Type Approval (R.C.S.131 Cat.A.H2) has been granted.

4. "METALPACK" PAPER TUBULARS with CERAMIC END SEALS New external construction gives

complete protection against moisture in 100% humidity at 100°C.

#### 5. H.V. TUBULAR CERAMICS

For Pulse Feeders in Radar equipment and Line Time bases in T.V. receivers. Ranges available for 1 kV. to 10 kV. D.C. working. Capacities from 10 to 620 pF.

## 6. HIGH VOLTAGE PAPER SMOOTHING CONDENSERS TYPE C.P. 561

For 25 kV. E.H.T. smoothing in large screen T.V. receivers. Absence of metal at "hot end" prevents corona losses.

#### 7. HIGH TEMPERATURE ELECTROLYTICS Ability to work at 85°C. without voltage de-rating. Characteristics low leakage current and high ripple rating.





## THE TELEGRAPH CONDENSER COMPANY LIMITED

RADIO DIVISION: NORTH ACTON · LONDON · W.3 · Telephone: ACORN 0061 (9 lines)

ELECTRONIC ENGINEERING

JUNE 1953





## ... as shown at the R.E.C.M.F. EXHIBITION IN LONDON.



Obviously, there is no room in this announcement for full technical data on these various Condensers.

We shall be happy to supply complete details of any range in which you are particularly interested, or to advise you of the types most suited to your particular requirements.

## 8. SUB - MINIATURE ELECTROLYTICS for HEARING AIDS & TRANSISTORS

Smallest Electrolytic ever made. Two sizes available—6  $\mu$ F for 3 v. D.C. working and 8  $\mu$ F for 6 v. D.C. working.

## 9. Hi-K CERAMIC DISC

For T.V. de-coupling and spark suppression. Two sizes—10 mm. and 20 mm. diam. Capacity range .001  $\mu F$  to .02  $\mu F$ .

## 10. IMPROVED "PICOPACK" ELECTROLYTICS

Smaller sizes and higher ratings are now offered. Voltage range extended to 450 v. D.C. at  $70^{\circ}$ C. Entirely new range for 85°C. working.

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**JUNE 1953** 

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## SPECIFICATIONS

Ture	LOAD	INPUT	INPUT CURRENT		OUTPUT	NO-LOAD	NET PRICI
IVPE	RATING	VOLTAGE	RATED	MAXIMUM	VOLTAGE	Loss	£ s. d.
50-A	5 kva.	115 v.	40 a.	45 a.	0-135 v.	65 watts	44 18 6
50-B	7 kva.	230/115 v.	20 a.	31 a.	0-270 v.	90 watts	44 18 6
	Loup		SERIES "	100" Variacs		Notorn	NET
TYPE	RATINO	INPUT	Cu	RRENT	Output	Loss	PRIC
		VOLTAGE	RATED	MAXIMUM	VOLTAGE		£ s.
100-K	2000 va.	115	15 a.	17.5 m.	0-115	20 watts	17 17
100-KM	2000 va.	115	15 a.	17.5 a.	0-115	20 watts	18 12
100-1.	2000 va.	230/115	8 8. ,	9 a.	0-230	25 watts	17 17
100-LM	2000 va.	230/115	8 a.	9 a.	0-230	25 watts	18 12
100-Q	2000 va.,	115	15 a.	17.5 a.	0-135	20 watts	18 9
100-QM	2000 va.	115	15 a.	17.5 a.	0-135	20 watts	19 4
100-R	2000 va.	230/115	8 a.	9 a.	0-270	30 watts	18 9
100-RM	2000 va.	230/115	8 a.	9.8.	0-270	30 watts	19 4
100-LH	1200 va.	480/240	2 a.	2.5 a.	0-480	25 watts	21 15
500-L O	1450 va.	180	8 a.	9 a.	0-180	25 watts	17 17
2000-K +	1000 va.	125	8 s.	9 s.	0-125	25 warea	17 17



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200-CM 200-CU }	860 va.	115 v.	5 a.	7.5 a.	0-135 v.	15 watts	7 17 6 6 15 0	
200-CMH 200-CUH }	580 va.	230 v. 115 v.	2 m. 0.5 m.	2.5 a. 2.5 a.	0-270 v. 0-270 v.	20 watts 20 watts	9 15 0 8 5 9	



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$a = \sqrt{a^2 + b^2}$	
= a bos a + b Sin a	POST THIS COUPON
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Overall lengths (mm	480	370		
Sensitivity	X	700	400	
$mm/V \times VA_8$	Y	700	350	
Y Capacity		<b>4</b> μμf	<b>4</b> μμ <b>f</b>	
Heater Volts		6.3	6.3	
VA <sub>3</sub> Max. KV	***	5	5	
VA1 Max. KV		2	2	
VA <sub>2</sub> (VA <sub>3</sub> =2KV) vol	300	300		
Vg for cut-off (max.)	-100	-100		
Drive for 20µa 1B	20v	20v		



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JUNE 1953



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# Electronic Engineering

Incorporating ELECTRONICS, TELEVISION and SHORT WAVE WORLD Managing Editor, H. G. Foster, M.Sc., M.I.E.E.

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### STABLE OUTPUT VOLTAGE OVER RANGE 1-40mA.

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Burning	current			•••			•••	l to	40mA
Increme	ntal resista	nce at 201	m <b>A (ap</b> j	prox.)	***	***	•••	* • •	<b>350</b> Ω
Tempera	ature coeffi	cient of b	urning v	oltage	e at 20m	A (app	rox.)	<b>_2.7</b> n	ıV/°C
Ambien	t temperati	ure limits					-55	to +	90°C
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MULLARD LTD., COMMUNICATIONS & INDUSTRIAL VALVE DEPT., CENTURY HOUSE, SHAFTESBURY AVENUE, LONDON, W.C.2 MVT132



Vol. XXV

JUNE 1953

No. 304

### Commentary

THE thoughts of millions of Her Majesty's subjects will inevitably be turned towards Westminster Abbey on June 2 when Her Majesty will be crowned Elizabeth the Second, by the Grace of God, of the United Kingdom of Great Britain and Northern Ireland and of Her other Realms and Territories, Queen, Head of the Commonwealth, Defender of the Faith.

A privileged few, no more than 7 000, by reason of their high office, will be present in the Abbey itself to witness thus solemn act of accession and dedication, while countless thousands will line the streets of London along the Coronation Route to acclaim our new Queen. But by the modern miracles of radio and television many millions at home and overseas will be brought to London to join in the ceremonies on this great day.

There have, of course, been many occasions in the past, notably the Royal broadcasts at Christmas, when the four corners of the earth have been linked to this country by radio, but never before on such an immense scale. The Coronation broadcast will, in fact, represent the greatest link up in history since nation first began to speak to nation, for over 200 narrators in London will provide commentaries throughout the day in some forty different languages.

It will also be a unique day, too. as far as the BBC Television Service is concerned, for although it is difficult to forecast an accurate figure, it is estimated that the British audience of viewers watching the scenes in the Abbey and along the Royal Route will not be less than twenty million. The problem of relaying direct across the Atlantic has not

yet been solved, but elaborate arrangements have been made to fly telefilm recordings across to Canada by R.A.F. Canberras within a few hours for transmission over the Canadian and American networks. The great achievement from the technical point of view is that viewers in France, Belgium, Holland and Germany will be joining British viewers by means of the temporary television relay network which has just been set up for the occasion.

This network is part of a larger scheme to establish a permanent network across Europe which will eventually

extend to the Scandinavian countries and across Switzerland into Italy. There are several European countries operating television services, but public interest has been at a low ebb largely owing to the poor standard of the programmes due to the sheer inability of the authorities to afford anything better. It has been realized for some time that one method of raising the standard would be in the exchange of programmes between neighbouring countries, and considerable attention has been given to this aspect in the development of the post-war European television systems. On a higher plane, too, it has been recognized that such exchanges would bring a wider culture to the viewer and go a long way towards promoting mutual goodwill and understanding among the nations.

The initial technical difficulties arising from the different television standards adopted by the European countries have to a large extent been overcome by the development of the convertor unit and the first exchange between countries with different systems took place last July when programmes originating in Paris were simultaneously transmitted in France and Great Britain.

In view of the world wide interest in the British Coronation, it was only to be expected that, while the time is not ripe for the building of a permanent network, the European television authorities would be anxious to relay the television programme of the Coronation to their countries. Some months have been spent by the BBC in discussing the arrangements with their European colleagues and plans for a temporary network were prepared early in the year. Progress in this direction was undoubtedly stimulated by

the BBC who offered the whole of the Coronation programme to the European network without charge.

The day is not far distant when a permanent network will stretch right across Europe from Scandinavia to Spain and it may yet be possible to bridge the Atlantic and join the American Continent. Viewers on Coronation Day, wherever they may be, might reflect on the words of William Haley Sir who described television as "an open window looking out on the world from every cottage in the land."

### The Editor and Staff of Electronic

The

Engineering send their loyal greetings to Her Majesty on the occasion of the Coronation on June 2

Coronation

LONG LIVE THE QUEEN!

### A Transmitter for Production Testing of Television Receivers

By J. M. Silberstein\*, Dipl.Ing., A.M.I.E.E.

A television test transmitter is described in which a comprehensive pattern is distributed to testing points simultaneously on all BBC channels. Pulses of appropriate frequencies, locked to the mains, are synthesized directly into the pattern, sync and suppression waveforms. Sync and pattern waveforms are separately impressed on 17-Mc/s. H.F. channels are formed by modulation with the necessary carriers using germanium double balanced modulators. Amplifier tuning is staggered for vestigial sideband transmission. Hybrid transformers join the vision and sound channels. Pulse and pattern monitors, and synchroscope for details of the television waveform are also described.

In mass-production of television receivers it is not possible to use, for testing purposes, signals transmitted by the BBC since the time of transmission is limited to a comparatively short period within the industrial working hours, the reception of these signals may not be good enough for testing, and only one channel is available in any area. A mass-production organization must have its own source of television signals, working all day and transmitting on all channels.

The signals provided by the test transmitter must be basically similar to the BBC television signals. The strength of the signals should be comparable with that in a good reception area, vestigial sideband transmission should be used, proper ratio of sync and video signals must be maintained, the relation between the sound and vision signals must approximate that met in actual working conditions.

Video modulation is practically the only feature of the transmitter which is left to the decision of the test gear engineer. Several basic solutions are possible and used. A camera and a film can be used, if a live picture is desired, or a camera and a picture or pattern card continuously scanned. These are very expensive solutions. A simpler method of producing a stationary picture is to use a high voltage cathode-ray tube as a flying spot scanner and a photo-multiplier tube or a photocell and a thermionic amplifier as a pick-up device. A transparent picture placed against the face of the cathode-ray tube can be made to produce a satisfactory picture because of the varying light intensity transmitted to the pick-up tube through the transparency.

A better method, of course, is to interpose a lens and focus an image of the scanning raster on to the transparency, as in this way better definition can be achieved. A different and probably more popular method makes use of a monoscope<sup>1</sup> which is a device incorporating a cathode beam-forming gun, similar to that in a cathode-ray tube, and a prepared target plate over which the beam is made to scan. The target plate has a picture or pattern printed or otherwise impressed on it, the ink used having a different secondary emission ratio as compared with the metal plate.

The replacement cost of the tubes for the above methods of stationary picture production has to be taken into consideration when the transmitters are running for the whole of the working week. Auxiliary equipment is needed in every case to provide essential synchronizing and suppressing waveforms, scanning and power supplies.

A properly selected geometric pattern is probably the best means of rapidly testing television receivers, and a

\* Advance Components Ltd., formerly Sobell Industries Ltd.

synthetic method of generating it from component waveforms is, in the long run, the most economical way of obtaining it. A pattern gives more quickly an insight into the operation of the tested receiver. Setting of the limits is much easier when a geometric pattern is produced on the screen, than when the tester has to watch a moving or even a still picture of natural, i.e. highly irregular design. The element of judgment is to a large extent replaced by objective criteria.



Fig. 1. Pattern used in the test transmitter

For normal production testing the pattern does not need to be very complicated. A simple pattern used in the test transmitter described, shown in Fig. 1, is sufficient to check all the main features which have to be watched in production, although it may not be adequate for development work on a new model. It permits, in conjunction with the controls of the tested receiver, a check of the picture size and of the aspect ratio; of the linearity in both directions, by observing the distances between the bars in different portions of the screen; of response, by observing the dots which correspond to high frequency components of the waveform; of time-base, by observing the clearness of the definition of the edges of bars; of interlacing, by observing the distances between the lines of dots; etc.

Besides all usual requirements of central testing equipment regarding its reliability and ease of maintenance, the test transmitter has to meet a specific condition which is

of paramount importance. The transmitter must remain locked, not rigidly, but in a "spongy" manner, to the mains even in the most adverse conditions. The mains frequency varies to a great extent and only an exceptionally sound locking circuit can guarantee the absence of breakdowns due to this cause. As all pattern generator designers and users well know, this is not a condition which can be met with an over-simplified circuit, however cleverly it has been designed. No expense in valves and components should be grudged to obtain the best results in this particular respect.

A simplified functional diagram of the test transmitter is shown in Fig. 2. The transmitter consists of the following main component parts:—

1. Frequency generator in which pulses of all frequencies needed for subsequent stages are derived from a single fundamental frequency. Whenever feasible, pulses are formed directly of the required shape and duration. The

required to shift the vision and sound signals into correct (BBC) frequency bands.

8. Channel vision modulators and final amplifiers bringing vision signals in the correct channels in the frequency spectrum.

9. Channel sound modulators and final amplifiers bringing sound signals (BBC broadcast programme) in the correct frequency bands to form five television sound channels.

10. Junction circuits combining sound and vision channels to be transmitted together over coaxial cables to the testing positions.

11. Monitoring arrangements comprising meters for carrier and signal voltages, a pulse monitor, a waveform display unit, a pattern display unit and a loudspeaker unit.

#### Frequency Generator

The analysis of the pattern (Fig. 1) gives directly the 'frequency of the pulses producing the vertical and horizon-



Fig. 2. The test transmitter

master oscillator is tied to the mains frequency through the locking circuit.

2. Dots generator for the high frequency components of the video waveform. These are generated at time intervals related to other components of the waveform.

3. Pattern synthesizer which joins together all the parts of the video waveform corresponding to the pattern selected.

4. Synchronizing signals generator (sync generator) in which the waveform is formed containing frame pulses, line and half-line pulses. This waveform is then delayed by about  $0.5\mu$ sec in a delay network.

5. Pattern suppressor in which those parts of the pattern waveform are suppressed which correspond to the periods of the flyback, horizontal and vertical.

6. Intermediate frequency modulator in which sync and pattern waveforms modulate an intermediate frequency.

7. Carrier oscillators producing the intermediate and channel frequencies (not the BBC channel frequencies)

tal bars. Vertical bars occur six times during one line (the sixth bar is not visible because it occurs during the horizontal flyback and is therefore suppressed), hence the repetition frequency of the pulses must be  $6 \times \text{line scan}$  frequency, i.e. 60 750c/s. The duration of the pulses must be about  $4\mu\text{sec}$  corresponding to the 1 : 3 ratio of the thickness of the bars and of the distance between the bars.

Horizontal bars occur 9 times during each frame (one bar is suppressed for the vertical flyback), hence the frequency required is 9 x frame scan frequency, i.e. 450c/s. The duration of these pulses is about  $250\mu$ sec corresponding to the thickness of  $2\frac{1}{2}$  lines on each frame.

Frequencies required for the sync signals are: 50c/s for the frame sync pulses, 10 125c/s for the line sync pulses and 20 250c/s for the half-line pulses. All frequencies are derived from one master oscillator,

All frequencies are derived from one master oscillator, its nominal frequency being 60750c/s. Actual frequency is identical with the nominal value only when the mains frequency is exactly 50c/s, and it changes in proportion to the mains frequency. A simplified schematic of the master oscillator and of the locking circuit is shown in Fig. 3. The oscillator is of the negative resistance type using a pentagrid valve. Its cathode, grid 1 and grid 2, form a reactance of value dependent on the potential of grid 1. As this reactance is connected across the oscillator tank circuit it can be used for automatic frequency adjustment.

The sine-wave produced in the oscillator is converted into positive going  $4\mu$ sec pulses in the pulse shaper and amplifier stage. The amplifier is required as fairly large pulses with a moderate source impedance are needed for subsequent operations. A single stroke multivibrator (univibrator) is used for pulse shaping in this as in many other cases throughout the transmitter.

A chain of frequency dividers follows the master oscillator, each producing positive going pulses of approximately 50V peak-to-peak. Nominal values of frequencies used in the chain are: 60 750-20 250-4 050-450-50c/s.

Pulses obtained from the last divider control the condition of a four diode bridge<sup>2</sup> across the other diagonal of which is applied a voltage derived directly from the mains. The value of the potential on the control grid of the reactance part of the oscillator valve depends on the phase relationship between the gating pulses from the 50c/s divider and the mains voltage. Whenever one of the two frequencies begins to change, the potential on the control grid changes and the oscillator frequency is brought



to agree with the mains frequency. The action is based on the changing instantaneous value of mains voltage occurring at the moment when the pulse from the frequency divider makes the diodes conducting.

A low-pass filter is connected between the output of the discriminator and the control grid to prevent a 50c/s modulation of the master oscillator.

The principle of operation of frequency dividers used<sup>2</sup> is illustrated in Fig. 4. At the input of the divider there is a two-stage amplifier. Two stages are necessary to preserve the correct polarity of the incoming pulses, otherwise the circuit would be triggered from the trailing instead of the leading edge of the incoming pulses. This would be equivalent to a delay between the originating and the generated pulse.

At each incoming pulse a large voltage is developed in the anode circuit of  $V_2$ . This voltage charges two capacitors separated by a diode. When the pulse ends,  $C_1$ discharges through another diode and  $C_2$  remains charged. After this process has been repeated several times,  $C_2$  has accumulated enough charge to make the triode  $V_3$  conductive, and this triggers the multivibrator causing a sudden rise of potential at the anode of  $V_3$ . This type of frequency divider, although rather expen-

This type of frequency divider, although rather expensive in valves by comparison with other, more popular types, has a number of attractive features. Capacitor  $C_2$  charges at a decreasing rate at each incoming pulse, but provided the last charging step is definitely higher than



#### Fig. 4. The frequency divider

the uncertainty limits of the triode to be triggered, the operation of the divider is quite decided. For a higher division ratio a higher amplitude of the charging pulses is required. Actually, a division ratio of 9 can be obtained without difficulty, and higher division ratios could be obtained if required.

The duration of the output pulses depends solely on the parameters of the multivibrator and can be adjusted within wide limits. The multivibrator is designed to work at a much lower repetition frequency than is forced on it by the triggering action. The divider is independent of the frequency within fairly

The divider is independent of the frequency within fairly wide limits. This is of particular value, because with the mains frequency varying from 48 to 50.5c/s many of the more economical dividers are prone to fail when working in a long divider chain. With the type of divider used it is possible to adjust all the dividers in the chain with the master oscillator running free and then just switch the locking on. If the adjustment was done correctly, the locking circuit will work whatever the difference of frequencies, within common-sense limits. Under normal conditions of supply voltage variations the locking circuit described can be expected to work unfailingly between 47 and 51.5c/s which provides a liberal safety margin.

#### **Dots Generator**

The vision signal produced by the test transmitter must contain some high frequency components to enable a rapid check of the frequency response of the tested receiver. These components are provided in the form of dots appearing on every fifth line. A switch serves to select very fine dots (2.7Mc/s), coarser dots (1Mc/s) or thin vertical lines, i.e. dots appearing on all scanning lines. It is desirable that the dots should not appear on the screen simultaneously with the vertical bars. A pulsed high frequency generator is used, shown in Fig. 5.

The oscillatory circuit of a Hartley type oscillator is connected in the cathode of a triode pulsed by 60750c/s pulses (vertical bars). When the triode is conducting, i.e. in the time intervals of vertical bars, oscillations can not occur because the oscillatory circuit is damped by the impedance of the triode. When the positive going pulse



ends. oscillations start immediately and continue until the beginning of the next 60 750c/s pulse. The oscillator is started afresh every time and this guarantees the recurrence of all the same effects every time. The dots can therefore remain stationary on the screen of the tested receiver, although their frequency is not tied

to the master oscillator. In one of the possible patterns the dots are to appear on every fifth line. To stop them on other

lines a pentode gate is used. 2 025c/s pulses are applied to the suppressor grid of the gating pentode which transmits the high frequency components applied to its control grid only during the time when the suppressor grid is driven positive. Positive going 2 025c/s pulses last for about 100 $\mu$ sec and are delayed by comparison to other pulses in the system by about 20 $\mu$ sec. This delay is obtained in the frequency divider 10 125-2 025c/s by omitting one stage of the input amplifier and thus starting the outgoing pulses from the trailing edges of the originating pulses. The delay is necessary to ensure that the dots start at the beginning of the horizontal scan line and not towards its end as would have happened otherwise.

To facilitate the gating action without introducing very high amplitudes of the gating pulses, the cathode of the gating valve is kept at a fairly high positive potential, whereas the suppressor grid is brought to earth via a resistor. Normally the pentode is nearly cut off and only a small additional negative pulse provided by the A.C. coupling of the gating pulses is required to complete the effect.

#### Synchronizing Signal Synthesizer

Component waveforms of the vision signal, i.e. the pattern and the sync waveforms, are formed from appropriate simple pulses in circuits called synthesizers. Of the two, the sync synthesizer is more interesting and will be described in detail.

The BBC sync waveform consists of line pulses and frame pulses interrupted by half-line pulses. Line and frame pulses are negative going, half-line pulses. Line and frame pulses are negative going, half-line pulses positive going. Television receivers respond to negative going pulses. To preserve the correct timing of triggering of the line time-base during the period of the frame pulse, it is necessary to shift line pulses by comparison with halfline pulses so that the period of time between the leading edge of the line pulse and the trailing edge of the half-line pulse equals 0 or a half-period of the line time-base. In the test transmitter described line pulses are derived from half-line pulses and their leading edges coincide. The first function of the sync synthesizer is therefore

The first function of the sync synthesizer is therefore to delay line pulses by the duration of the half-line pulses. This is done in a coincidence circuit. Line pulses of 20µsec duration are applied to the control grid of a pentode and half-line pulses of half duration and double repetition frequency are applied to the suppressor grid. Half-line pulses are first amplified in a triode amplifier and their polarity reversed. A simplified circuit is shown in Fig. 6. It is obvious that anode current can flow in the pentode only when the potential of the suppressor grid is positive and the potential of the control grid is not highly negative. This coincidence can occur only during the second half of the line pulse and in this period a large negative going pulse occurs at the anode of the valve. This pulse can be used as line sync pulse because it meets the requirements stated previously.

All the component pulses are now available for the



synthesis of the sync waveform. A simple addition of the pulses would not be satisfactory as half-line pulses must not occur except during frame pulses.

The process of addition is performed in a mixer comprising three triodes, each having one type of pulse applied to its control grid. The output is taken from the common cathode impedance. The polarity of half-line pulses is the reverse of the polarity of the remaining pulses. Fig. 7 shows component pulses, A,B,C, and the direct result of their addition, waveform D. The latter contains in its middle part the exact form of the required waveform. The problem consists therefore of finding some simple means of removing the upper and the lower parts of waveform D.

The two triode amplifiers following the mixer are used as a waveform shaper. As their cathodes are earthed, they have a tendency to suppress all positive parts of the incoming waveform. The first triode suppresses the upper part of the waveform p, the second, the polarity having been reversed, suppresses the lower part of the waveform. The result of this operation is shown in Fig. 7 as waveform E. The operation does not call for any particularly critical adjustment of the relative values of the amplitudes of component pulses. One variable resistor is provided in the common cathode load of the mixer and this is quite sufficient.

The BBC television waveform has sync signals delayed

#### Fig. 7. Formation of the synchronizing waveform



by  $0.5\mu$ sec to enable the receiver to recover from the peak white signal before the sync signal begins. In the case of the pattern as described above the delay is actually not necessary, but as some other types of video modulation were considered for introduction at some later date, the delay has been provided. To obtain it without upsetting the all-important rise-time of the sync pulses, a low-pass filter is used designed for a cut-off frequency of 5Mc/s. It is a ladder network containing inductors as the series arms and capacitors as shunt arms. The characteristic impedance is low and a cathode-follower stage is necessary before the filter.

#### I.F. Vision Modulator

The pattern waveform synthesized in a similar manner to the sync waveform must first pass through the suppressor stage before it can be used for modulation. The suppressor consists of a suppression waveform synthesizer and of a pentode gate. The pattern waveform leaving this stage is not changed fundamentally, but it is extinguished for the duration of the flyback, horizontal and vertical.



Fig. 8. The I.F. vision modulator

It is a distinctive feature of the test transmitter described that the two waveforms: pattern and sync are not combined before the modulation (i.e. on the video basis), but are used separately for modulating two consecutive stages of the modulator.

A simplified schematic of the I.F. modulator is presented in Fig. 8. Suppressor grid modulation is used in both stages. In the first stage the sync waveform, brought to an amplitude of nearly 100V in the preceding amplifier, is applied to the suppressor grid of a pentode which has the I.F. carrier voltage (0.25V, 17.7Mc/s) applied to its control grid. The result of this process is an I.F. signal of constant amplitude, except at the periods corresponding to the sync pulses (not half-line pulses) when the amplitude drops to 0. This signal is applied to the control grid of the second modulating pentode. The gain of this stage is controlled by the pattern waveform applied to the suppressor grid. When the pattern pulse (positive going) is present, the gain is about three times higher than in the periods between pattern pulses. The exact ratio of the change of the gain is controlled by the variable resistor in the cathode which makes it possible to find the correct working point. An I.F. signal appears at the output of the second modulating stage which has zero amplitude in the sync pulse periods, full amplitude in the pattern pulse periods and 30 per cent amplitude in all other periods (black level).

amplitude in all other periods (black level). This system of modulation is characterized by the desirable feature of independence of the sync ratio (ratio of the black level to the peak white level) of the picture content, at least in the first approximation. In the test

transmitter it makes it possible to change the pattern without readjustment of the sync ratio. With a different type of video modulation, i.e. live picture, the advantages would be still greater. In the test transmitter further advantages are gained by the fact that all video stages operate at two conditions only, whereas in the case of adding the sync and pattern waveforms on the video basis the amplifier preceding the modulator would have to work at three different conditions; zero signal, black level, full signal.

The intermediate frequency stage of modulation presents an easy way of distributing the vision signal to the five channel modulators. If a direct system of modulating channel carriers was used, a large amplifier would be required for video signals, working at low impedance level for the sake of coaxial cables used for distribution. The distribution at I.F. frequency is incomparably easier. Further advantage is gained when a part of the networks required for vestigial sideband transmission is transferred from channel modulators into the I.F. modulator with the resultant saving in the total number of tuned circuits in the transmitter.

#### H.F. Transmitter

Carrier frequencies used in the channel equipment are equal to the BBC channel carrier frequencies minus the test transmitter I.F., so that correct vision carriers are obtained at the output of the channel modulators when the upper sidebands are filtered out. Double balanced bridge modulators are used, with germanium crystal rectifiers and transformers wound on spiral tape cores. The difference between the carrier and modulating frequencies is fairly great in all cases and the elimination of unwanted modulation products does not present any serious difficulties. Tuned circuits in the amplifier stages following the modulator are tuned in such a manner that, in conjunction with the tuning of the I.F. stage, an overall frequency response is obtained approaching the BBC vestigial sideband transmission system.

Sound channel equipment is similar to the vision equipment. The intermediate frequency for the sound modulation has been chosen 3.5Mc/s lower than for the vision, i.e. 14.2Mc/s, and this makes it possible to use the same carrier frequencies for both the vision and the sound equipment, as 3.5Mc/s is the distance between sound and vision carriers on all BBC channels. The common usage of the same carrier sources for the sound and the vision imposes rather stringent requirements as regards leakage in the modulators, to avoid vision breaking into sound and vice versa.

The I.F. sound channel is obtained from the so-called programme generator reproducing an original BBC programme (Light or Home) on several frequencies, for testing radio receivers on all frequency bands. In the programme generator the original transmission is not demodulated to the audio stage. Various frequency bands are obtained by modulating the BBC transmission with appropriate carriers. By this means full modulation depth is preserved without difficulty.

All carrier frequencies used in the test transmitter are crystal controlled.

It is desirable to have the monitoring circuits, pattern display and loudspeaker units, as close to the final output of the test transmitter as possible, otherwise the faultless operation of the transmitter may be in doubt even when the monitors are producing normal effects. To avoid the mutual interference of the sound and vision signals which would seriously interfere with the monitoring, hybrid transformers are used for joining sound and vision signals to be sent out on common coaxial cables. At a loss of just over 3db in the main transmission path, a separation of the two signal sources is obtained amounting to at least 30db. Hybrid transformers are wound on spiral mumetal tape cores 0-001in. thick.

Nowhere in the H.F. circuits are really high levels used. All connexions are made at 75 ohms impedance by means of coaxial cables. Through these precautions it was found possible to avoid elaborate screening of the H.F. parts of the test transmitter.

#### **Monitoring Facilities**

For maintenance purposes the test transmitter is amply provided with monitoring facilities. All I.F. and H.F. channels, vision and sound, have at their outputs rectifier voltmeters to keep the transmission levels under constant control. Similar voltmeters are provided for the carrier outlets. An audio amplifier and loudspeaker can be switched to the output of every sound channel where rectifiers are provided for the purpose. In a similar manner a video amplifier and vision receiver, complete with sync separator and time-bases, is provided for monitoring the outputs of all vision channels, the detected vision signals being taken from germanium rectifiers connected immediately before the junction hybrids.

A simple double-beam oscilloscope is provided for rapid check of the frequency dividers belonging to the standard frequency generator. Double beam operation is particularly important in this case as the best way of checking the dividers is to display simultaneously the originating and the



Fig. 9. The pulse monitor

generated pulses. The number of pulses occurring between two consecutive pulses of the slower rate shows directly the division ratio, all imperfections of adjustment are shown by the untidy appearance of the generated pulses, or even, in the most drastic case, by the impossibility of keeping one of the traces in lock.

The block schematic of the pulse monitor is shown in Fig. 9. The time-base is locked to the faster of the two displayed signals because the checking operation has obviously to start at the higher frequency end of the chain (with locking circuit out of action). A beam switching system is used for displaying the two waveforms. The beam is switched from one trace to the other at a frequency equal to half the frequency of the time-base. Both waveforms pass through amplifiers to the Y-plates of the cathode-ray tube, but at any moment only one of the amplifiers is actually working and the other is switched off by the action of the switching waveform applied to its suppressor grid. Blanking pulses are provided to suppress all display during the return stroke.

Normally the pulse monitor is kept in the position where it shows on one trace 50c/s pulses produced in the test transmitter and on the other trace 50c/s mains sine-wave. The two waveforms do not move with respect to each other when the transmitter is in lock.

Another specialized oscilloscope (synchroscope) is provided to display all details of the television waveform and in particular to keep control over the sync ratio. This is an oscilloscope with a triggered time-base. The triggering frequency is 25c/s as the whole of the television waveform is repeated at this rate. The speed of the trace in the horizontal direction has to be such that between 10 and 20

lines are displayed. The speed of the time-base must be completely independent of the triggering pulses. The beginning of the sweep should be determined by the time location of the triggering pulse. Fig. 10 illustrates the principles of the operation of the synchroscope.

It is an essential requirement that at every moment the triggering frequency must be identical with the picture frequency, otherwise two consecutive displays would not exactly coincide on the screen, the display would be blurred or subject to spasmodic jerks. This is not an easy requirement to meet. Dr. Sturley<sup>3</sup> described the difficulties met while building a similar apparatus. The solution adopted by Dr. Sturley is practically the same as the one derived independently for the test transmitter synchroscope. Broadly speaking, it consists of abandoning all efforts of making use of the mains frequency and developing instead a 50c/s sine-wave from pulses produced in the transmitter. In spite of the fact that the test transmitter is locked to the mains, the instantaneous value of the frame pulse frequency may differ from 50c/s mains sufficiently to cause the jitter. One of the reasons for it may be that in the long divider chain hunting processes occur while the frequency is controlled and adjusted. Observation showed that the jitter was particularly severe when the mains frequency was far off its nominal value, and so the locking was somehow more difficult, although this could not be noticed by any other means.

To display any part of the television waveform a phase



Fig. 18. The synchroscope

shifting network is provided following the 50c/s generator. It covers more than 360° which is enough to show one complete frame and the beginning of the next frame—the only period when the two frames differ.

The phase shifted 50c/s sine-wave is first squared and then applied to the input of the 50-25c/s divider producing very short pulses used for triggering the time-base in which a capacitor is charged via a constant current pentode circuit. A constant speed of the electron beam is thus obtained for the forward stroke. The speed of the beam, i.e. the number of the scan lines displayed, can be regulated within sufficiently wide limits.

The input signal to the synchroscope is the detected output of any I.F. or channel vision output amplifier. The connexion has to be done at low impedance level and so the signal level is very low. An amplifier is therefore provided in the synchroscope with a uniform gain in the video range; it consists of three pentode stages.

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## The Relative Merits of A.C. and D.C. as a Signal Source in Analogue Computors

By K. H. Simpkin\*, A.M.I.E.E., A.M.Brit.I.R.E.

The author has attempted, in this article, to marshal the factors upon which the choice of A.C. or D.C. as a signal source in analogue computors is based: the influence of this decision in the case of a large computor, such as a flight simulator, is considered in detail, both from the point of view of equipment design and of the overall accuracy and sensitivity of the computor. The design of power supplies, signal sources, and amplifiers, and the relative complexity of simulating mathematical operations are examined in relation to both A.C. and D.C. systems.

While no dogmatic decision (which would apply universally) can be attempted, the author does collate, in a usable form, the points which must be considered before a choice of signal can be made.

WHEREVER the design of analogue computors is discussed, there is usually considerable argument when the decision has to be made between an A.C. and a D.C. computor. Both methods have their devotees, but an examination of the problem will show that neither system is so overwhelming in its advantages as to remain in universal use.

The author has had occasion, in the past, to take part in such arguments and presents this article in the hope that, while not providing a final decision in itself, it may present the "pros and cons" in a condensed form, and thereby assist readers, faced with a similar problem, to make the decision themselves.

The type of computor envisaged in this article is of formidable size and complexity, such as might be used in a flight simulator, for example. The power supplies and the numbers of basic units called for are large, and any economy which might result from the choice of a particular system can have a marked effect on the price, running costs and the reliability of the computor.

In view of the wide nature of the subject, it is proposed to examine the merits of A.C. and D.C. under separate sections, each devoted to a particular aspect of the computor.

#### The Signal Supply

In a D.C. system, any amplitude variation of the signal supply results in a direct error proportional to that variation. In a large computor, this supply may be of the order of several kilowatts and is normally of low voltage, in order to reduce the dissipation in the resistive computing elements. The problem of supplying highly stabilized power at potentials of less than 100 volts inevitably involves the use of low efficiency stabilizing elements, which increase the complexity of the computor and the total power required. A rotary generator is normally called for, which is not easily obtainable these days, together with the means for stabilizing its output voltage and removing ripple, the components of which may lie in the useful band of the computor if any eccentricity is present in the rotor of the generator. Furthermore, since both positive and negative values of functions are normally handled in a computor, supplies of both polarity are required, one of which must be stabilized from the other in order to guard against zero errors.

\* Air Trainers Ltd.

In compensation, however, no radiation is experienced with direct current and provided that the source impedance of the supply is kept low for reasons of stability, there is no necessity for screening the main supply to prevent crossmodulation, since there are no frequency components of the signal of sufficiently high order to be subject- to radiation.

In an A.C. system, error is produced by both amplitude and frequency changes in the signal voltage. It is essential, therefore, that both be stabilized. A rotary alternator is necessary in view of the high power required, together with the associated equipment for stabilizing the voltage and the frequency and for removing the unwanted harmonic content of the supply, since distortion of the signal voltage results in further errors in the computations. Filters for removing harmonics while passing several kilowatts of the fundamental frequency involve substantial losses and contribute to the power requirement of the whole.

When a suitable A.C. supply has been produced, it must be protected from adulteration by external signals and prevented from inducing unwanted voltages in the various circuits of the computor, since any such voltages contribute directly to the errors of the system, and may even result in instability of the computor. Extensive screening is therefore necessary in most cases, a factor which increases the cost of the computor and may add to the difficulties of experimental work undertaken during its development.

experimental work undertaken during its development. In the opinion of the author, unless it can be shown that the signal power requirement is markedly less, in the case under consideration, when an A.C. supply is used, the D.C. system has the advantage from the point of view of signal supplies.

#### Amplifier Power Supplies

In a D.C. amplifier, the working conditions of the valves depend upon the difference in potential of a positive and a negative H.T. supply voltage. Since these voltages may each be of the order of 200 times the bias voltage required, it will be seen that a high order of stability is desirable in these two supplies in order to prevent the valves being swung from cut-off to saturation by small changes in the supply line voltages. Furthermore, variations in these supplies result in direct errors in the output of the amplifiers, although this effect can be combatted by negative feedback, provided they are not of such magnitude as to affect the forward gain of the amplifier materially. Heater supplies tend to be complex in a D.C. amplifier.

however, since, in a high gain amplifier, a stabilized D.C. heater supply is generally necessary in the first stage to minimize drift and it may be necessary to provide several sources of heater supply when the valve cathodes in the various stages of the amplifier are working at appreciably different D.C. potentials.

There being no D.C. equivalent of the transformer, an amplifier is required whenever the signal needs to be changed in impedance or reversed in sign. The number of amplifiers required to solve a particular problem tends therefore to be greater in the D.C. system and, as a corollary, the H.T. power requirement also. This tendency is accentuated by the fact that, owing to the loss in coupling circuits in D.C. amplifiers, the number of valves in each amplifier is large and the power efficiency correspondingly low. The responsibility for providing these supplies at low

The responsibility for providing these supplies at low impedance rests wholly with the power supply designer, since the signal voltages are also at D.C. and additional decoupling at each amplifier cannot be relied upon. The stability of the whole computor rests therefore on the ability to produce and maintain a low source impedance in these supplies.

The H.T. supplies required for an A.C. system are far less complex. In general, the power output and gain for a given valve line-up is greater in the A.C. amplifier. Negative feedback enables a fair degree of amplitude variation and hum to be tolerated in the H.T. supply and, since the valves are isolated from one another at D.C., a single positive supply is needed.

It may be necessary to provide an additional heater supply to minimize hum pick-up in the first stages of high gain amplifiers, but the main heater supply can be common to all other stages.

Changes of impedance or polarity in the signal can be accomplished with a transformer, with the result that the number of amplifiers and hence the supply drain associated with them tends to be less in the case of the A.C. computor. The economic importance of power supplies in a large computor leads the author to believe that the arguments put forward in this section are strongly in favour of an A.C. system.

#### **Computing Errors**

Errors which are common to both types of computor, such as non-linearity in potentiometers, will not be considered.

In a D.C. computor, errors normally arise due to unwanted D.C. voltages in the signal channels. These result from supply variations and drift in the computing amplifiers. The former have already been considered in the previous section. Drift is a serious problem, for its removal requires the careful design of input stages, the selection of valves, and, in some cases, a more positive attack in the addition of a drift correction system, involving the use of a separate amplifier. Such amplifiers demand additional power from the system and, since their function is not primary in nature but arises solely due to the deficiencies of the computing amplifiers, their presence is a serious argument against the use of D.C. for computing purposes. Some relief from drift can be obtained by the use of high signal levels, however, since a direct improvement in signal-to-noise ratio results.

Errors peculiar to an A.C. system arise from two major causes. These are the presence of stray pick-up voltages of signal frequency in the computing circuits and the introduction of phase errors by the various components of the computor. Prevention of the former lies in extensive screening as has been mentioned before. Phase errors arise usually from the use of iron circuits, such as transformers and rotary resolvers, or as a result of variations in the supply frequency of the signal. The reduction of these phase errors involves the addition of numerous correction networks to the signal circuits and the application of large

amounts of negative feedback over the non-linear elements in the amplifiers. Phase correction networks are generally sensitive of adjustment and may become misaligned should the signal frequency change slightly. Furthermore, owing to the spread of iron characteristics and the difficulties of maintaining accurate gap settings in the manufacture of transformers and rotary resolvers, it is often necessary to align, individually, the correction circuits of each amplifier. in a production batch. Feedback over transformers demands careful control over the loop characteristic and results in the addition of further correction networks to the feedback chain. It appears therefore that phase errors are a serious drawback to an A.C. system and that much of the simplicity claimed for A.C. may be lost in the correction of these errors. No improvement can be sought in raising the level of the signal voltages, since errors due to phase are independent of the amplitude of the signal. The signal/error ratio can be improved only by the actual reduction of the phase error itself.

In view of the advance made in recent years in the design of stable D.C. amplifiers, the author is of the opinion that a D.C. system today is less subject to errors than an A.C. system of equivalent complexity.

#### Arithmetic Addition and Subtraction

Simple addition and subtraction is normally based, in an analogue computor, on the properties of Kirchoff's laws. These involve either the addition of voltages in a common mesh or the summation of currents at a common node. Since, however, p.c. voltages cannot be produced in an isolated form, except by the use of separate power supplies, the addition of voltages in a mesh is impracticable in a D.C. system. The addition of currents at a node is suitable to D.c. techniques, but involves a considerable loss of signal when the number of terms to be added is large. An amplifier is almost always necessary for adding, and drift and gain limit the range of addition available, since the output current at the grid node is always at a lower level than any other of the input currents.

Subtraction involves the use of negative signal supplies locked to the positive ones or else the introduction of a sign reversal by means of a buffer amplifier.

Since the transformer allows impedance changes without loss of power and the production of isolated signal voltages, it is possible, in an A.C. system, to use either mesh or nodal addition at will. The former system is preferable. however, since, by isolating each signal with a cathodefollower amplifier having a very low output impedance and summing into a high impedance load, it is possible to add voltages in a mesh with negligible power loss.

An A.C. system, however, adds voltages vectorially and it is therefore necessary to bring all the signal voltages into common phase before addition is possible.

Subtraction is no more complex than addition when a suitable transformer is available to reverse the polarity of one of the signals concerned.

The addition and subtraction of angles may be achieved equally well in either system, when the angles are available as shaft positions, and do not therefore come within the scope of this article.

Since some form of phase correction is required to make use of an A.C. system at all, the author is of the opinion that arithmetic addition and subtraction is simpler and less demanding in power when an A.C. system is used, than in the case of a D.C. system.

#### **Multiplication**

The multiplication of two voltages may be achieved in either of two ways. These comprise a group of methods under the generic title of "electronic systems" and the electro-mechanical method of multiplication. The latter involves the application of one of the voltages to a potentiometer, the wiper of which is positioned in accordance

with the other voltage. There is little to choose in this latter system between A.C. and D.C. as a signal source. Potentiometers used with A.C. signals may require phase correction, but, on the other hand, there is a greater range of small A.C. motors available, which are suitable for instrument servo use.

Since non-linear impedance methods rely on the stability of the characteristic used being high, to be of value in computing techniques, there is an advantage in the use of an A.C. signal. The A.C. characteristics of valves, for example, are more stable and reproducible than the D.C. characteristics, which are less easy to control during manufacture. This criticism does not apply, however, to digital and sampling systems, in which an on-off characteristic is used, but these systems are outside the scope of this article.

It appears, therefore, that, since electronic methods of computation are inaccurate in both systems in the present state of the art, there is little to choose from the point of view of multiplication and division between the two systems.

#### **Trigonometrical** Solutions

There are no suitable resolving or compounding devices available which operate at D c. Sine-cosine potentiometers suffer from a lack of resolution, unless built in an inconveniently large physical size, and Desynn resolvers are, at best, inaccurate devices for use in computing. It is generally necessary, therefore, to re-cast the problem, in a D.C. system, by the use of co-ordinate geometry for example, in such a way as to avoid the need for the resolution or the compounding of angles.

The design of A.C. resolvers has, however, reached a high stage of development; rotary transformer devices are available with high degrees of accuracy and resolution. These however do suffer from the lack of a definite zero, especially when there are traces of quadrature signal present. It is generally recognized, therefore, that for the solution of purely trigonometrical problems the A.C. computor has outstanding advantages over the D.C. model.

#### **Integration and Differentiation**

Electro-mechanical integration, by means of a servo motor whose speed is maintained proportional to the input voltage, is equally achievable by either system. Electronic integration, by the use of a Miller capacitor integrator, is essentially a D.C. device and has no equivalent in an A.C. system. In a computor, the high speed available in such an integrator is often necessary to the stability of the system, and the lack of an equivalent unit at A.C. is a very serious drawback to the purely A.C. system.

Differentiation by similar feedback methods is possible in the D.C. computor, but considerable care is needed in the design of the amplifiers to avoid instability and the maximum rate must be limited to avoid excessive deterioration of the signal-noise ratio.

Thus, in both these applications of the calculus, and indeed in any application calling for the shaping of the voltage function, a D.C. signal is essential.

#### Servo-Mechanisms

The behaviour of a high performance servo-mechanism is controlled by the motor amplifier characteristic and the performance of the motor.

A.C. servo motors have reached a high stage of development and, for a given size of motor, outweigh the D.C. motor of the same general characteristics in speed-range, acceleration and torque. The majority of D.C. motors available today have been designed for low voltage applications and, as a result, they require stabilized low voltage and high current supplies. These are inefficient to generate as a separate supply, while the provision of individual carbon pile regulators for each armature may become prohibitive in a large computor. Furthermore, these low voltage motors

require high field currents to drive them, and these currents have to be supplied by the control amplifier. It is a wellknown fact, however, that valve amplifiers are more suitable to the production of high voltage, low current outputs, than high currents at low voltages. The presence of brushes in a D.C. motor is an additional drawback, as they restrict the life of the motor, affect the slow running characteristics and produce spark interference, which can cause serious trouble with associated equipment such as the radio aids equipment in a flight simulator.

The control of the amplifier characteristic is another case of voltage shaping, and a D.C. signal is essential for obtaining the transient gain terms necessary for high grade servo performance.

Thus, the advantages seem to be shared equally between the two systems.

#### Amplifiers

The major design problem of D.C. amplifiers is drift in the input stage. Several methods of drift correction have been evolved in recent years, however, and in less critical cases careful design of the first stage and valve selection results in an amplifier with a useable drift figure.

Positive and negative H.T. line voltages are necessary and, due to the heavy loss in D.C. couplings, the gain per stage tends to be low. This is not always such a disadvantage, however, since the D.C. couplings can be made part of the loop correction networks. The loop design is simpler in a D.C. amplifier than in an A.C. one, since the phase shift is due only to capacitive shunting.

High stability resistors are necessary to maintain stable D.C. working conditions on the valves, and some form of zero-setting is normally essential. Furthermore the requirement of an earthy input and output, when the signal is zero, usually results in further loss of gain.

Hum is no great problem, except in so far as it must not be allowed to overload any stage or be rectified in non-linear impedances. Heavy negative feedback will reduce hum and other unwanted interference, except drift, to a harmless level. There is an inherent disadvantage, however, in D.C. amplifiers, that an odd number of stages is required for a feedback amplifier, which usually results in a minimum of three valves being used.

Linearity is not so critical as in an A.C. system, where non-linearity results in the production of harmonics of the input signal which are difficult to remove without causing phase shift of the fundamental.

The efficiency of D.C. amplifiers is low in the output stages, since the valve must be worked at the midpoint of its characteristic to give linear swings in both polarities. From a more general point of view, component replacement tends to be more critical in D.C. than in A.C. amplifiers.

In A.C. amplifiers, owing to the necessity of using transformers, considerable ingenuity is required in the design of the feedback loop. More valves are often required than might be apparent in order to obtain sufficient phase control to avoid oscillation. Furthermore, the phase shaping net-works usually involve the use of tuned elements which are critical of adjustment.

Valves have no p.c. connexion to one another, and no loss is involved in coupling networks, so that the gain per stage is high, although limited by phase shift due to shunting capacitances.

Highly stable elements are necessary only in the feedback loop, for a high degree of feedback minimizes variations in the forward gain elements. The efficiency of output stages can be made high by using class-B operation, provided that distortion is reduced to an acceptable level by large amounts of negative feedback. The additional gain then necessary can be obtained from valves of low current drain.

The problem of component replacement is almost entirely obviated by heavy negative feedback. In view of the fact that the design of D.C. amplifiers has

been the subject of so much research in recent years, the choice between the two types of amplifier rests largely upon the "know-how" of the designer. Neither amplifier involves insuperable design problems, although, in the opinion of the author, the balance of advantage all round lies with the A.C. amplifier.

#### Non-Linear Functional Representation

Since there is, at the time of writing, no method of representing non-linear functions with adequate accuracy other than by the use of graded potentiometers, which are common to both systems, the subject will not be pursued further.

#### **Response** Time

On the reasonable assumption that A.C. amplifiers for computor use are designed as a low-pass amplifier with a low frequency cut-off, it will be obvious that the bandwidth necessary for a given speed of response in a D.C. amplifier will be less than half that required in an A.C. one. With the high degrees of feedback in use in computing amplifiers, 80db or more, this means that the loop response must be controlled for an additional octave, in the A.C. case, above the asymptotic limit of the D.C. amplifier. These asymptotic frequencies lie in the region of 50 to 100kc/s, where control is difficult and any alleviation in the frequency limit is of great help to the designer. The reduced bandwidth of the D.C. amplifier also results in a theoretical doubling of the gain/bandwidth factor; if the phase control design will allow it, a reduction in the number of stages required for a given application may thereby be achieved. It appears, therefore, that the D.C. amplifier has the advantage in respect of response time.

#### The General Wiring and Layout of a Large Computor

In a D.C. computor, wiring tends to be simpler, since screening can be dispensed with except where necessary to avoid instability. The frequency limit of the signals is sufficiently low as to make crosstalk unlikely and stability can be maintained, in some cases, by restricting the amplifier bandwidths. Some care is necessary, however, in applying this cure as an afterthought.

At A.C. the problem can be formidable, although it is no worse than any other crosstalk consideration in the communications field, except that high and low level voltages tend to be routed in close proximity to one another. During development, however, more care is necessary in bench layout since, at 400c/s for example, pick-up in high gain amplifiers without adequate shielding can result in false accuracy figures.

In the experience of the author, neither system has over-

#### MICROPOWDER MAGNETS

MORE than 20 years ago the General Electric Co., Ltd., began to produce iron dust components for electrical work. It has now developed a technique for making magnets -equal to any made from the finest cobalt steel—from microscopically fine iron dust made by a special process. The new magnets will be known commercially as Gecalloy magnets, and will be produced by the Salford Electrical Instrument Co., Ltd., a G.E.C. subsidiary. Powder metallurgy is, of course, not new. Experiment

Co., Ltd., a C.E.C. subsidiary. Powder metallurgy is, of course, not new. Experiment showed, however, that the relatively coarse powder which was suitable for the core of an inductor was the exact opposite of what was required for a permanent magnet. The particles used are only 100 times the diameter of an atom of iron and  $1/1000^{\text{th}}$  the diameter of the finest radio iron powder hitherto manufactured.

powder intherto manufacturea. To produce the minute particles of pure iron was in itself difficult but in addition there were certain problems of handling to be overcome. Many fine metallic powders when loose are pyrophoric, i.e., they ignite on coming into contact with air; and therefore a method of handling had to be

whelming advantages, although the D.C. system is less likely to give trouble if previous experience in the layout of such systems is limited.

#### **Combined Techniques**

It will be obvious that a system taking advantage of the best of both worlds would be an ideal choice. Such a system requires means for the conversion of A.C. signals to D.C. and vice versa. Unfortunately, such devices as do exist for that purpose have disadvantages of their own.

Conversion from an A.C. signal to a D.C. one can be achieved by the use of various forms of phase sensitive rectifier circuits. Most of these result in a certain amount of ripple in the resultant D.C. signal. Those that do not do so require to work into an infinite load impedance and are successful provided that the requisite conditions of operation are complied with.

Conversion from a D.C. signal to an A.C. one involves the use of a modulator. Valve modulators are unreliable in operation, since they introduce effects, analogous to drift, due to variations in contact potential of the rectifiers. A more reliable choice of modulator is the vibrating relay. As far as is known by the author, however, there is only one relay manufactured in quantity in this country which is suitable for such purposes, and that is both difficult to obtain and expensive for extensive use in a commercial computor. It is essential that the relay used be capable of maintaining a constant mark-space ratio indefinitely, without the need for constant maintenance, and of vibrating at frequencies up to the order of 1000c/s.

The major drawback of mixed systems, however, is the need of suitable power supplies of both types. This may well complicate the design to a greater extent than if the deficiencies of one or the other were tolerated.

#### Conclusions

The merits of A.C. and D.C. as a signal source in analogue computors have been investigated in detail. It will be apparent that there is no serious argument in favour of either as a whole. The choice depends on two main factors. These are, firstly, the type of problem to be computed and hence the units of equipment necessary for the task, and, secondly, the experience and personal judgement of the designer, which is, after all, the deciding criterion in any complex project.

#### Acknowledgments

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evolved which would remove this danger.

The results of exhaustive tests have proved conclusively that all the requirements of a good magnet have been met by the new process.

Until now the different steels used for magnets have all had one property in common. They are extremely hard and therefore the fabrication methods are limited to casting and

therefore the faorication methods are limited to casting and grinding Gecalloy micropowder magnets can be made—after the powder has been produced—in a great variety of shapes by the use of special press tools and power presses. Another factor of great importance is that, strength for strength, a Gecalloy magnet only weighs approximately half as much as a conventional steel magnet, which will commend the micropowder magnet for use in the aircraft industry and for all mobile equipment for all mobile equipment.

This material should also prove of use for such applications as loudspeaker fields and television focusing magnets while the lack of eddy currents will commend its use in inductive devices. It is also worthy of note that this process will allow the production of magnets made only from pure iron obtainable in Britain.

### Printed Circuits and Miniaturization

By P. Eisler\*, Dr. Ing.

This article assumes that at present printed circuits for use in electronic equipment are largely for the purpose of wiring and shielding. The printing of actual components is not dealt with. The bearing of foldable, three-dimensional circuits on the miniaturization of electronic equipment is emphasized, and the main design features are pointed out. As long as most components are not printed integrally with the wiring these features will retain at least some of their validity.

**RINTED** circuits are being used to an ever-increasing extent in electronic equipment of many kinds. Usually they have the shape of small wafers and are made of plastic, glass or some ceramic material "decorated", by means of a printing or stamping process, with a metallic pattern of silver or copper lines. Often such wafers form the coverplates of cast resin blocks inside which groups of small components are clustered. The terminals of the components, whether single or potted into sub-units, project through perforations in the wafers and can be solderjoined to the metallic lines of the wafers, for example, by means of a dip soldering bath. Other mechanization—or automatic assembly—procedures, are also possible with such wafers. In the interstices between the metallic lines some types of resistors can be deposited. Flat capacitors or tape resistors can be soldered on to the wafers, and spirals of a few turns may be provided as printed inductors.

The above briefly characterizes the well-known common type of printed circuit, irrespective of the particular "printing" method used. It has won the support of some of the U.S. and U.K. governmental agencies looking after service requirements, and has begun to be appreciated though not yet on a large scale—by private industry. The defence interest in these methods of construction of electronic equipment is centred around two features, which also have an important bearing on civilian production. These are:—

- (a) the miniaturization of equipment made possible by these production techniques.
- (b) the saving of labour, tooling and change-over time in mass production and servicing of electronic equipment under war-time conditions.

Even if some of the possibilities listed below can be realized only when the printed circuit technique is much more developed than now, the interest taken by defence ministries in this technique is readily understood, so selfevident is the importance of saving weight, space and material in airborne and portable military equipment, not to mention the desirability of avoiding or reducing to a minimum the mobilization for electronic production purposes of large masses of workpeople in times of emergency. Consider, too, the advantages of freeing production from tooling operations and tooling delays, since the changeover from an old design to a new one can be effected by simply making a new drawing and printing plate. Also to be borne in mind is the way in which maintenance can be greatly facilitated, and even in part dispensed with.

Two main features of printed circuits, namely, miniaturization and labour-saving. become conflicting however, when the miniaturization is taken to a stage where the components are packed so closely (with due regard to the intrinsic electrical, thermal or mechanical limitations) that they become difficult to assemble, to test, or to exchange, in short, when the unit becomes a "watchmaker's job." As far as the bulk of electronic equipment is concerned, this "watchmaker's" stage is reached long before we come near to the limit of the miniaturization of components or of the definition of printing, (i.e., the degree of distinctness in the outline of the printed pattern).

The chances afforded by the art of printing and by all the other techniques capable of making for a much greater degree of miniaturization (of components particularly) must remain unexploited, as with the wafer form of construction the result would be assemblies so fiddling, so over-crowded and inaccessible, so unrepairable, etc., that costs would soar to exorbitant heights. Smallness alone is not objectionable from a labour cost standpoint, but over-crowding and the comparative inaccessibility of numerous joints of small components are.

It is usually found that in spite of the above mentioned limitation set by the requirements of accessibility, the reduction in size and cost which can be effected within these limits is by no means negligible, as when, for example, a wireless receiver previously made according to orthodox methods is being redesigned for mass production by the wafer type of printed circuits.

Nevertheless the field in which miniaturization by the wafer type of printed circuit can be applied is restricted because both production and maintenance of any piece of electronic equipment of which it forms a basic part demand a comparatively large amount of space for the purpose of accessibility. The demand for further miniaturization is insatiable, and it appears certain that any marked progress in miniaturization, apart from the ever-increasing extension of the printing technique to the components themselves as integral parts of the printed circuit, calls for a fundamentally different approach—a return in fact to first principles.

#### The "Three-Dimensional" Printed Circuit

Such an approach is exemplified by the idea of the three-dimensional printed circuit, or "folded" circuit, as it is sometimes called. This circuit consists of a piece of thin sheet or tape which, in the finished equipment has a three-dimensional configuration so that the printed conductors spread over the space in all three directions. They can therefore assume the functions of wires, thus obviating the necessity of going to the length of designing equipment with all conductors in one plane. They can form practically any three-dimensional network or pattern, and can conform to the shape and position of most component terminals. They also permit a direct connexion with component terminals owing to their flexibility and relatively large contact surface.

Instead of a three-dimensional pattern of lines which we have in the case of wire conductors, we have a threedimensional arrangement of surface patterns. Our elements are *surfaces* in space, not linear conductors.

In the following only one major feature of the idea of three-dimensional printed circuits is described, namely, the scope it affords regarding the miniaturization of a piece of electronic equipment as a whole as distinct from the miniaturization of its actual components.

<sup>\*</sup> Technograph Printed Circuits Ltd.

#### Packing into the Minimum Space

There are obviously several factors which set limits to the theoretically minimum space which any piece of electronic equipment needs, for example, the number and size of the electrically active components, their shape and the electrical, magnetic and thermal characteristics which determine their minimum theoretical distance from each other, the grouping of the components according to functional or maintenance requirements, etc.

If it were possible to achieve a packing factor of unity, that is, to arrange all active components in a space where one component would be so close to its neighbour (subject to the limitations set by the above-mentioned physical characteristics) as to leave no space between, and furthermore, if it were possible to eliminate all electrically nonactive parts, for example, all parts or components having only a mechanical and structural function, then with the given components the ultimate degree of miniaturization conceivable would be reached.

Obviously, such packing and virtually complete elimination of other than electrically active components are for all practical purposes unattainable, but adoption of the three-dimensional printed circuit does permit some approximation to this goal.

Taking the printed circuit itself, it consists usually of a very thin, flexible sheet of insulating material such as paper or (plastic) film carrying on one or both surfaces a flexible metallic pattern. It is usually made by the foil technique<sup>1</sup>, as foil conductors are superior to others besides excelling them in flexibility, but any technique capable of producing highly flexible suitable circuits could in principle at least be used. Thus, vacuum and electrodeposited metallic circuits on thin films are quite feasible where this type of conductor is not open to objection for other reasons. But techniques which form the conductor from a metallic powder with or without a liquid medium and consolidate the powder by heat or pressure or both, are counter-indicated mainly because of the brittle nature of the conductor thus produced.

#### Working in the Flat

The flexible sheet constituting the metallic pattern with its insulating support which has a three-dimensional shape in the complete equipment is produced in the flat, and as a flat sheet it is first given creases where it is to be subsequently bent. As a flat sheet it is also stretched, thickened, plated, punched, slit, cut, insulated by coverings and pre-pared for solder or other joints. While it is still a flat sheet, single components or groups of components, assembled or produced as a sub-unit elsewhere, are affixed to it either by hand or by automatic means. Only after all this is done in the flat is the circuit distorted into the third dimension. It is "folded" into the configuration the designer has foreseen and for which he has developed the flat pattern. If the sheet is folded at an angle of 180° the folding is said to be superimposed. The folding itself may be referred to as a fold or bend, and even as a winding if the angle of folding exceeds 180°. Depending on the radius of curvature the folding can be either smooth or sharp. There are some well-known regular folding configurations such as the concertina type folding and various folded box constructions which are fairly easily produced by automatic machines. Obviously the designer should be aware of these facilities. However, practically any spatial configuration of the originally flat circuit is possible, and the designer has therefore complete freedom

to bring the components into any arrangement relative to each other and to situate them anywhere within the space available. The circuit can conform to any position; it becomes "three-dimensional."

In order to make the best use of this feature for the achievement of the maximum degree of miniaturization in the production of any piece of electronic equipment, the first problem facing the designer is how to arrange the components of the unit into as closely packed a grouping as is compatible with functional requirements, and also with such overall considerations of design as fitting to other units. Within these limits, but without any thought for the moment as to accessibility, mechanical structure, etc., he has to cram the components into the minimum amount of space.—(At this point some such picture as that of sardines packed in a tin ought to be at the back of his mind!)

Having settled on a desirable spatial arrangement of the components, the three-dimensional configuration of a thin sheet containing the electrical connexion of the components is devised, and this conception is then transferred to the drawing board where it appears flattened out (developed) into one (enlarged) plane design. Crossconnexions are taken care of by splitting the pattern into two superimposed ones, for example, into two patterns on opposite faces of the sheet. When the components are drawn in or placed on the flat development of the patterned sheet at their respective connexion points, it at once becomes obvious that many components are distanced further apart from each other on the flat development than they were in space on the patterned sheet in its three-dimensional configuration.

#### Spacing by Folding

This is an essential feature of the "folded" circuit and what has been demonstrated by observation, when develop ing the three-dimensional circuit into a plane, becomes a consciously planned feature of the design aimed at conveniently separating during their assembly components which will be in very close proximity in the finished equipment. Any two points of the three-dimensional printed circuit sheet between which it is folded (bent around or otherwise distorted) owing to its configuration, become wider apart from each other when the sheet is developed into one plane.

Even where the sheet in the finished equipment has no marked spatial configuration folding of the sheet on to itself is a readily adopted design feature. The threedimensional nature of the circuit pattern in such cases is only nominal. But the effect of the folding-up of the sheet is very marked, resulting, as it does, in the closest possible packing together of components which were so far apart from each other on the flat sheet prior to folding that they could be assembled, tested, exchanged and worked on in perfect ease both by manual and mechanical means.

It must also be remembered that in order to allow for the development of automatic means of component insertion into a circuit, as well as to guard against damage by heat when soldering for purposes of maintenance, and to save labour, very tight packing of components during assembly should be avoided. The folded circuit enables this precaution to be taken, and yet still permits the components to be grouped into the minimum space.

In some cases the circuit is designed on a moulding or inside a casting. Here again, it is often advantageous to execute it as a flat development, carrying out as far as possible all operations on the circuit while it is in the flat state. Bonding it to the moulding, or packing it in, would be one of the last operations. Thus, work in a restricted space with poor accessibility may be avoided and a smaller moulding or casting may often be feasible.

From the above the principle underlying the claim of the foldable circuit technique to make possible a second degree of miniaturization in the production of electronic equipment can readily be seen, but except for "potted" assemblies two important points need clearing up. before much use can be made of this principle.

#### Maintenance of Miniaturized Equipment

The first point concerns maintenance. There is a school of thought which tends to deny the importance of maintenance considerations for mass-produced equipment. It favours the building-up of equipment from a few "potted" throw-away units, and quite a number of arguments based on economics or military logistics are advanced in support of this view. Undoubtedly, this viewpoint is amply justified in certain instances, but in general it cannot be accepted by either the industry or the user of the equipment for reasons not open to question by the designer. The problem of maintenance (and mechanical support) must therefore in most cases be faced, and the easy wayout offered by potting cannot be resorted to.

Between accessibility and the ease of maintenance there is a direct relation. This is one of the reasons why miniaturization creates special maintenance difficulties. ment. It is quite clear that a thin flexible sheet like the one used for a three-dimensional printed circuit cannot reliably support any but the smallest components. While wafer type circuit made of a comparatively strong а material like bakelite laminate or a ceramic can be used as a small chassis and support a fairly wide variety of components, no mechanical load or stress can be put on the foldable circuit. A separate mechanical structure has to be provided, which far from using the circuit as a support must actually support the circuit as well as the component. In some cases a chassis or a moulding provides this structure, the circuit being affixed to it in such a way that both the circuit and the components joined to the circuit are properly supported (see Figs. 1 and 2). This, however, is not a preferred method of design. Ideally only electrically active and indispensable components should be present in a piece of miniaturized equipment. A mechanical structure, such as a chassis or a moulding, does not belong to this category and should consequently be eliminated from the equipment wherever possible. Obviously many benefits would accrue to both producer and user from a realization of this aim.



Fig. 1. An experimental amplifier

The three-dimensional circuit is not subject to the limit on the extent of miniaturization normally imposed by accessibility. At first glance, therefore, maintenance of such a circuit would seem to be impossible, and not merely just difficult as in the case of the wafer type of circuit. However, this conclusion is unwarranted. For the folded circuit can be unfolded again—and herein lies one of its biggest advantages. A sheet in a three-dimensional con-figuration developed out of a flat sheet can be brought back into the flat again for the purpose of exchanging components, effecting repairs, or making tests of all kinds. Of course, the circuit will in the first place have had to be designed with this in mind. For re-conversion of the three-dimensional configuration into a flat pattern practically the whole circuit must be complete in the flat, with none, or at most, only a few of the joints having to be undone when unfolding the circuit from the spatial shape into the flat. Nor must any major mechanical dismantling be involved in unfolding. It is obviously not possible to test all performance characteristics while the circuit is stretched out in the flat, but as folding and unfolding can be repeated in quick succession, a facility and ease of maintenance is created which is often not obtainable even with ordinary non-miniaturized equipment. In respect of accessibility, maintenance is thus made just as easy as production.

#### **Mechanical Structure**

The second point calling for special attention is the provision of a sound mechanical structure for the equip-



Fig. 2. Another view of Fig. 1

The function of the mechanical structure is to support. Shielding is already provided by the circuit and by the way in which components or sub-assemblies have been constructed. What, then, can provide the required support if a separate structural device is not desired? We have already found that the circuit sheet cannot do so. There remain only the components themselves. Clearly only the larger components can be expected to provide such support. Here, in fact, is the solution to the problem.

There are larger components (coil or tuning assemblies in metal containers, valves, transformers, instruments, controls, etc.) in most types of electronic equipment, and even where all the components are very tiny larger mechanical units of several smaller components can be easily created by packaging some as sub-assemblies. This packaging, too, can be done with the aid of a smaller printed circuit of either the foldable or wafer type. It can also be done by potting (casting) or cementing the small components into a unit, or by shrinking a plastic or elastomeric bag or tube over the group of components in question.

One of the most promising but as yet untried techniques in this direction appears to be the use of a foldable printed circuit made as (or for use with) self-adhesive tape to which the small components are joined while the tape is stretched out in the flat. The tape is then folded (wound) in such a way that its adhesive layers become superimposed and thus stuck together in an easily handled unit with the small components enclosed and reasonably sealed up between layers of adhesive tape, but with all required connexions leading from the tape.

#### Interlocking Components

The problem of the main assembly of the type of electronic equipment under consideration can therefore be taken as the problem of assembling the larger components and the sub-assembled units of small components into a mechanically self-supporting stable structure. The complication of the tiny components is overcome by grouping and fixing them into a larger unit. To bind them together it is not too difficult to provide these larger components or units with interlocking features (plugs, sockets, tongues, grooves, keys, slots, springs, hooks, pins, bolts, holes, etc.) which are so arranged that they interlock when the components are brought together by folding-up of the flat sheet into a three-dimensional configuration. The larger parts thus lock themselves together into a structure of any desired rigidity depending on the location and type of the interlocking device used. They thus form themselves into a self-supporting combination and at the same time also support the printed circuit to which they are joined at their terminals.

Here we have another principal difference between the wafer type and the flexible, foldable printed circuit. Whereas the former usually acts as a support or chassis for components, the latter is usually supported by the components themselves. It is readily seen that the substitution for a separate mechanical structural device of interlocking of the components achieved by the act of folding, parallels the great advantage in ease of maintenance that the possibility of readily unfolding the circuit into the flat affords. The components become unlocked when the circuit is unfolded, and in this way the mechanical structure is temporarily dissolved.

The role of potting or casting in such construction work is limited to the embedding of groups of small components together with interlocking devices, thus transforming these groups into larger interlocking units. Usually such a unit is not designed to permit exchange of its small components, but in some cases such provision can be made, for example, by using a two- (or more) section cast featuring a parting line along which the components are placed, and preferably making at least one part of the cast compounds from a flexible rubber-like material.

Although the basic patents and original development work on the three-dimensional printed circuit technique described above date back some ten years, comparatively very few practical examples have so far been published for a variety of reasons. It is hoped that it will not be equally long before general recognition is awarded to the claim of this technique to effect a second degree of miniaturization of electronic equipment, and at the same time satisfy any desirable level and requirement of maintenance, unit construction, and exchangeability of components.

The figures illustrate some of the aspects described in the article. Figs. 1 and 2 show an experimental 30db amplifier of the Post Office, one of the earliest uses of a three-dimensional flexible circuit in connexion with a chassis. This circuit is not designed to be unfoldable again. A full description of the equipment for which the printed circuit was designed has been published<sup>2</sup>.

Figs. 3 to 6 illustrate various aspects of a design of a miniaturized 5-valve superhet "personal" wireless receiver. The size of the set is roughly that of a cigar case. and while this particular design would now be considered too complicated, it nevertheless includes most of the features of the interlocking concertina circuit described above which can be folded and unfolded. Fig. 3 shows the circuit-sheets stretched out in the flat. Fig. 4 and 5 are cross-sections of the equipment showing the three-dimensional configurations (folded concertina) of the circuit sheets, and Fig. 6 gives a view of the "large" interlocking mechanical parts, forming in this case a semi-rigid inter-

locking structure by means of just two pairs of tongues and grooves and one bow. A full description of the equipment has been published<sup>3</sup>, while some of the principles outlined in the article are the subject of a British Patent<sup>4</sup>.

#### Summary

Leaving aside for the purpose of this article the development of components produced as integral parts of a printed circuit, and taking the latter as providing merely interconnexions and shielding of components, the wafer



Fig. 3. The circuit sheets in the flat



Figs. 4 and 5. Cross-sections of the equipment showing the three-dimensional configurations

type of printed circuit makes possible miniaturization only to a degree compatible with the accessibility of components. The three-dimensional foldable circuit permits miniaturization to be taken a stage further by virtue of the following:

1. It is a thin flexible sheet occupying only a very small space owing to its thinness.

2. It can be folded (deformed) from the flat into practically any three-dimensional configuration, and thus permits of the packing of components into any space, like sardines in a tin, irrespective of accessibility.

3. It can be unfolded from its spatial shape, becoming a flat circuit again, thus permitting access to all its com-

1

0 0 0 no D Ъ 0

Fig. 6. The interlocking coil can assembly

ponents while in the flat state, both for assembly and maintenance purposes.

4. It enables the convenient distancing during assembly and maintenance of neighbouring components which are very close in space, as unfolding the circuit into the flat gets neighbouring points on opposite sides of a fold apart from each other.

5. It can dispense with many intermediary metal parts between circuit conductors and component-terminals as it can assume any shape and can contact terminals directly over a relatively large area.

#### **Special Quality Valves**

#### An Announcement by the B.V.A.

IT has been a matter of general knowledge that valve manu-facturers have supplied for many years special valves for such uses as Post Office repeaters. In addition they, have undertaken in the post-war years a large scale development to produce valves which will give trouble-free service under onerous conditions such as are encountered in communications and radar equipment in aircraft, warships, and armoured fighting vehicles where the apparatus is subjected to severe vibration and shock.

It has become the custom among those concerned with the development and use of such valves to describe them as "reliable" valves. The B.V.A. considers that such a term can be misleading in its implication in that it may convey the impression in some quarters that all the valves not specifically so designated are unreliable.

The valves sold by the B.V.A. members for use in ordinary domestic radio and television sets are designed expressly for this purpose. It is obvious that such domestic sets are not this purpose. It is obvious that such domestic sets are not subject to the severe conditions of vibration and shock men-tioned above and the valves used in them are entirely adequate on the score of mechanical strength and useful life. This is borne out by the satisfactory low level of field returns experi-enced over many years. It should also be noted that they are produced at a satisfactory economic price for the market concerned.

Thus the valves which have been developed for the onerous conditions mentioned in the earlier paragraph are in fact special quality valves designed for a variety of special applications, and as they are coming into more general use the B.V.A. intends to describe these valves in future as "Special Quality" valves.

The reason for the use of this term can best be appreciated from a knowledge of the technical definition of such valves which is as follows:

A "Special Quality" valve is a valve which has certain design and manufacturing features making it suitable for use under conditions different from or in excess of those experienced in normal radio or television receivers and when operated under

6. It (often) enables all or much of the mechanical structure to be dispensed with altogether by combining the folding scheme with means for interlocking of the larger components and of the units of smaller components, so that they themselves form a structure which supports both itself and the circuit.

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stated or agreed electrical or mechanical conditions it has an acceptable statistically determined expectation of life. The various electrical and mechanical operating conditions result in several classes of "Special Quality" valves among which are the following:

- Class 1: Valves to give particularly long lives under conditions where they are not subjected to appreciable mechanical shock-Examples: repeaters and computors.
  - 2: Valves to give normal lengths of life under moderate conditions of vibration and shock-Examples: equip-ment for the Armed Forces, civil aircraft, mobile communications and some industrial uses.
  - Valves to withstand particularly severe conditions of vibration and shock where comparatively short lives 3: can be tolerated-Examples: projectiles and guided weapons.
  - 4: Valves to give high electrical stability and normal lives under conditions where they are not subjected to appreciable vibration and shock-Example: D.C. amplifiers.

Note: These classes must not be regarded as mutually exclusive and combinations of the attributes inherent in each class may be incorporated in a particular type. It must be realized that "Special Quality" valves require

special design, special manufacture, special quality valves require stringent testing at all stages. The necessarily greater cost is justified because the failure of such valves in the equipments concerned can involve the failure in service at a critical moment

of a large fighting unit and loss of life. While these "Special Quality" valves are intended mainly While these "Special Quality" valves are intended mainly for Government service there is little doubt that in later years the ordinary domestic valve, good as it now is for its purpose, will be improved still further. The relations between the "Special Quality" valve and the normal commercial valve is similar to that between the special high priced racing auto-mobile and the normal domestic motor car. The analogy is almost perfect since it is well known that the reliability and efficiency of the standard British motor car progresses year by year in proportion to the efforts made by designers and manu-facturers to produce better and better racing cars. facturers to produce better and better racing cars.

## Blocking Oscillators

#### By K. G. Beauchamp\*, A.M.Brit.I.R.E.

The operation of electronic equipment, especially where associated with cathode-ray tubes, often necessitates the generation of non-sinusoidal waveforms of a pulse or sawtooth nature.

Many circuits exist for this purpose and perhaps the most used, and certainly the most economical in valves and components, is that of the hard valve blocking oscillator.

This can be considered as a limiting case of the squegging oscillator and is characterized by a short period of oscillation followed by a relatively long quiescent period.

Several forms of the circuit exist, and are extensively used in three different ways: (1) as a generator of voltage or current pulses, (2) as a generator of repetitive sawtooth current waveforms, (3) as a generator of repetitive sawtooth potential waveforms.

As the literature appertaining to the subject is so widely dispersed and in some of its aspects, nonexistent, it is the purpose of this article to examine these three cases in some detail and outline their mode of operation and factors that influence their practical design.

I n radar equipment, blocking oscillators find many applications, as frequency dividers, and as low impedance pulse generators for triggering and switching purposes.

Very large current pulses can be obtained from a low-mu triode of the 6SN7 type if the positive grid excursion is used, as can be seen from the curves shown in Fig. 1. Where the ratio of pulse length to repetition period is kept small in order that the dissipation rating of the valve is not exceeded, pulses of the order of 1 ampere are easily obtainable. Voltage pulses may be obtained either from the anode or cathode circuit, or alternatively a separate transformer winding may be used.



Two modes of operation are possible with the simple blocking oscillator shown in Fig. 2. In order to describe these, we may suppose the cycle to commence with a large negative charge at the grid sufficient to reduce the anode current to zero. The grid potential will begin to rise exponentially as the charge leaks away through R, towards the potential to which the grid is returned.

If this potential is greater than that at which the mutual conductance  $(g_m)$  of the valve would increase to such a value that the gain around the feedback loop exceeds unity, then regeneration will commence. The anode current  $(i_n)$  increases, with consequent drop in anode potential  $(v_n)$ , the reflected increase in grid potential causing a further increase in  $i_n$  and so on, the process being rapidly cumulative, until at length grid current flows. As peak cathode current is reached  $i_n$  begins to fall, the grid acquiring a large negative charge. A second quiescent period ensues and the cycle repeats once more. This mode is known as the "astable blocking oscillator".

It will be seen from the above that two time-constants

are involved, the larger one due to the grid recovery circuit RC, and the shorter one controlled by the feedback circuit.

Should the grid leak return potential be smaller than that necessary for regeneration, a monostable condition results, wherein an external control potential must be applied in order to produce the requisite current pulse. For this reason the mode is known as a "triggered blocking oscillator".

This is sometimes used in association with a delay line to produce a self-triggered arrangement. Here the generated pulse is passed through the delay line, the delayed output pulse being used to trigger the oscillator once more.

In this way the properties of the artificial line, which can be strictly controlled, may be used to regulate the pulse repetition frequency.

The design of the coupling transformer involves maintaining the H.F. and L.F. response such that the desired pulse shape is obtained. Low leakage inductance (i.e., good coupling factor between windings), small stray capacitance, and core losses are necessary to ensure a good H.F. response, while high permeability of the core material is desirable to improve response at the L.F. end.

Low frequency response of the transformer controls the pulse duration, providing that C is large enough to have negligible impedance within the pass-band. With a small C the transformer windings have less effect and pulse width is reduced with reduction of C. However, the resonant period of the transformer and associated stray capacitance must be short compared with the pulse duration, thus setting a limit to the minimum pulse width obtainable by this method.

In practice, circuits and transformers have been designed to give pulse widths of from 0.1 to  $25\mu$ sec.

#### Sawtooth Current Generation

A form of astable pulse generator such as is described above, has been used by several workers to control the charge and discharge of an inductive circuit, resulting in the formation of repetitive sawtooth current waveforms.

The earliest analysis of this type of circuit was given by Vecchaicci<sup>1</sup>, who showed that an over-coupled triode oscillator having the ratio of inductance to capacitance, L/C, greater than the square of the triode internal impedance,  $R_0^3$ , is capable of generating continuous oscillations of sawtooth current waveform.

Due to the finite and varying  $R_0$ , coupled with the resistance of the transformer windings, the current increases

<sup>\*</sup> General Electric Co., Ltd, Coventry.

exponentially rather than linearly with time during the scanning period.

As the sawtooth currents generated are usually employed to scan an electro-magnetically deflected cathode-ray tube for television and other purposes, a high degree of linearity is essential, and many attempts have been made and methods evolved to linearize this current.

One method is to include a gapped iron-cored choke in the circuit<sup>2</sup> where the inductance varies with the current flowing through it due to a partial saturation of the magnetic core.



where:  $L_0$  = total inductance of the circuit (including that of the choke and scanning coils with no direct current applied through windings).

R =total circuit resistance.

E = applied H.T. potential.

i = instantaneous value of current.

This principle has been elaborated by Taylor<sup>4</sup> who has produced several circuits using a gapped choke and using high impedance scanning coils in both triode and pentode circuits.

A typical circuit of this type is shown in Fig. 3, where it will be seen that the current through the deflector coils,  $L_{\rm D}$  and also the voltage fed back from anode to grid is modified by the inclusion of choke CH.

As the current through this choke increases, its inductance is decreased, permitting an additional increase of current to take place. This additional increment of current, above the normal rise due to the charging action of the



Fig. 5. Scanning coils connected in low impedance grid circuit

Fig. 6. Simplified circuit of Fig. 5

inductive circuit, approximates to the correction curve shown in Fig. 4, and linearizes the current waveform generated.

As with the pulse generators previously described, the repetition rate is controlled, in the absence of synchronization by the time-constant RC in the grid return circuit.

Synchronization can be effected by applying a positive pulse to the grid directly or through a separate winding on the transformer.

For television purposes, however, low impedance line

scanning coils are desirable. The reason for this is that the rate of change of scanning current di/dt is large and of the order 6 to  $10 \times 10^4$ A/sec, therefore the voltage pulse across these coils, v = L di/dt must be kept within the limits imposed by the nature of the coil insulation. Because of the peak voltage incurred and the large scanning currents required in the coils, certain modifications become necessary to the circuits previously described.

A form of blocking oscillator has been described by Malling<sup>5</sup> in which the coils are included in the low impedance grid circuit, the linearity during the scan period being adjusted by suitable variation of grid-cathode impedance. A high slope, large anode dissipation valve is used, with the anode connexion brought out separately and insulated for several thousand volts.

This circuit is shown in Fig. 5, and in order to understand fully, its mode of operation, which differs from the pulse generators already described, a simplified circuit is given in Fig. 6.

given in Fig. 6. The valve V has a low impedance during the scan period and the reflected impedance of, the scanning coils has a high ratio of inductance to resistance, L/R, so that the anode current rises exponentially. Regenerative feedback takes place, due to coupling between anode and grid by T, and soon grid current commences to flow.

Both anode and grid currents increase until cathode saturation results; further increase of grid current can then occur only at the expense of anode current and the latter falls. The effect is cumulative and the valve is rapidly cutoff by a large negative charge accumulating on the grid,  $g_1$ .



Fig. 7. Equivalent circuit of Fig. 6 Fig. 8. Anode current/voltage curves

The inductive circuit, together with its associated stray capacitance, forms an oscillatory circuit and resonates under the stimulus of the application of a sudden change in potential across it. After half a cycle of oscillation, the charge on  $g_1$  is dissipated and V commences to conduct once more, the cycle then repeats.

The difference between this mode of operation and that of an astable pulse generator is thus seen to be that with the former a long conducting period for V is permitted, followed by a short quiescent period during which the valve cathode current is cut-off by a large negative grid bias. With the pulse generator, however, the reverse is true; a short conducting period followed by a larger quiescent period.

The circuit of Fig. 6 is thus characterized by two unstable states which alternate with each other to form a continuous sawtooth current waveform; viz:

(1) A rigorous single sine wave oscillation of frequency  $\omega^2 = 1/L_pC$  followed by a period of heavy grid current in which:

(2) Exponential rise of current takes place.

The sine wave oscillations are determined entirely by the  $L_pP$  product, while the rise of current is governed by the length of the damped grid current period, which in turn is controlled by transformer design and partly by applied negative grid bias.

It will be seen, therefore, that the circuit may simply be represented by the equivalent circuit of Fig. 7, where the valve acts as a switch  $S_1$ ,  $S_2$ , both poles of which are

closed during the scanning period and open for the retrace period. The two impedances  $Z_a$  and  $Z_g$  representing anode and grid impedance of the valve.

From the  $i_0/e_a$  and  $i_g/e_a$  curves for a typical valve (Fig. 8) it is seen that while  $Z_a$  remains substantially constant,  $Z_g$  changes considerably over the working range.

Hence the grid current  $i_g$  does not exactly follow the exponential charge of anode current and as the scanning coils are included in the grid circuit, this effect modifies to a certain extent the waveform of current through the coils.



This is made clear in Fig. 9 where  $Z_g = \delta e_g / \delta i_g$  is shown to cause a variation in grid current over the valve conducting period, in a manner determined by its varying rate of change.

The adjustment of  $Z_{\rm g}$  is carried out, empirically by variation in the amount of energy fed back from the anode to grid circuit by alteration to the transformer turns ratio and applied negative bias. This will also be found to effect the frequency generated, and in practice, owing to the absence of a complete mathematical treatment for this time-base, the correct ratio and bias can be determined only by experiment.

The way in which the repetition frequency is controlled by the applied negative bias and hence R may be derived if we consider the scanning period as the D.C. charging period of  $L_p$ , i.e.,  $S_1$ ,  $S_2$  closed in Fig. 7. Then as the current increases, a negative bias is developed across R of Fig. 5, this limits the amount of  $g_1$  voltage excursion into the positive grid region and hence charging time. As the retrace time is fixed by  $L_pC$ , then the sawtooth current repetition frequency is controlled by variation of R.

Synchronization may be obtained by the application of a negative pulse to the grid a little before cathode saturation occurs, this will initiate the retrace by decrease of anode current.

Because of the long conduction period for the valve used in this arrangement, it must be of a generous power dissipation rating and capable of passing a large cathode current during a major part of the sawtooth cycle.

A heavy grid current pulse flows in the valve towards the end of the scanning period, a mean value of 40mA being quite common, hence care must be taken that grid emission does not take place.

During the retrace period the rapid changes in anode and grid current induce a potential in the grid winding of 1 to 2kV. Adequate insulation in the valve base and holder is thus necessary.

Despite the operational difficulties mentioned above, however, several manufacturers have in the past successfully incorporated this arrangement in commercial television receivers.

#### Sawtooth Voltage Generators

A fundamental defect of the type of line scanning circuits dealt with in the preceding section is that the valve is conducting over some 90 per cent of the scanning cycle. This leads to a very poor efficiency and for this reason is now seldom used in commercial television equipment.

An alternative arrangement is to use the "resonant

return" system, first proposed by Blumlein<sup>•</sup> and late, developed by Schade<sup>7</sup>, wherein part of the energy stored in the deflector coils during the retrace period is returned to the circuit as useful power. A full description of these circuits is given elsewhere<sup>7,6,9,10</sup>, but the particular aspect of their mechanism which deserves inclusion in this section is that a sawtooth waveform voltage generator is, in general, used to control these circuits. This is so often a blocking oscillator of the pulse kind considered earlier that it is proposed to devote some space to their consideration.

A blocking oscillator was first used as a hard valve timebase in 1923 by Appleton, Herd and Watson-Watt<sup>11</sup>. This followed the form of an astable pulse generator charging a capacitor by grid current through the valve and allowing the charge to leak away during the quiescent period (Fig. 10).

The repetition rate of the current pulse generated is such that the capacitor C never completely loses its total charge received during the valve conduction period. By allowing the capacitor to discharge to about 20 per cent of the potential to which the grid leak R is returned the exponential discharge curve approaches a linear rise of voltage with time.

This approximation to a linear rise is not always adequate, however, and several methods have been evolved to linearize the waveform generated. In the original patent a diode was used in place of the

In the original patent a diode was used in place of the resistor R, enabling a constant charging current to be obtained for C by using the cathode current saturation effect of the diode. Due to its  $i_n e_n$  characteristics a pentode may also be used in the same manner to keep the charging current constant and therefore the rise of potential across C linear with time.

Another method of improving the linearity is to use a correction circuit due to Hawkins<sup>13</sup>. This employs an auxiliary charging circuit across part of the charging capacitor C, which in this case consists of two capacitors  $C_1$  and  $C_2$  in series (Fig. 11).

At the commencement of the cycle  $C_1$  and  $C_2$  are charged rapidly in a time of t seconds.

The integrating circuit  $R_1$ ,  $C_3$  has a larger time-constant of the order T = 75t and thus receives very little charge during the period t seconds.  $C_1$  and  $C_2$  then commence to discharge slowly;  $C_3$  discharging exponentially, while  $C_1$  discharges into the lower impedance path  $R_1C_3$  and charges  $C_3$  until no potential difference exists between points A and B. When this condition is reached  $C_3$  begins to discharge through  $R_1$  and R.



Thus the potential difference across  $C_2$  is of exponential form (a) while that across  $C_3$  is parabolic in shape (b). The total waveform across  $C_2$  and  $C_3$  is therefore the sum of (a) and (b) and is represented as waveform (c). By making  $R_1$  variable, T can be adjusted until the summated waveform is a linear rise of potential with time (Fig. 12).

Returning to the mechanism by which the capacitor C receives its charge. During this period, when the value is conducting, the circuit behaves as a Hartley oscillator,

Fig. 13, the frequency of oscillation being controlled by the transformer inductance and stray capacitance in a manner determined by the relationship:

$$f = 1/2\pi \frac{1}{\sqrt{(L_p + L_s + 2M) C_s}} c/s \dots (2)$$

where  $L_p = primary$  inductance.

- $L_{\bullet} = secondary$
- M =mutual
- $C_{\bullet} = \text{stray capacitance.}$

For correct operation with minimum retrace time it is essential that the first positive half cycle, after the valve is cut off shall not be of sufficient amplitude to bring the valve within its grid base. For this reason a very low Q factor is necessary for the tuned circuit, if the circuit is not to continue to function as a class C sine wave oscillator, the anode potential never being reduced to zero for a sufficient length of time to cause the valve to be cut off.

To avoid this effect use may be made of the eddy



Fig. 13. Effect of transformer iosses on Hartiey oscillator



Fig. 14. Blocking oscillator (a) functioning normally and (b) with insufficient t.ans.ormer damping

current,<sup>1</sup> and hysteresis losses of the transformer core, which may be represented as a parallel resistor  $R_{\circ}$  across the tuned circuit of Fig. 13.

High losses in the iron core or alternatively external damping across the transformer windings are thus desirable with this form of blocking oscillator if the retrace time is not to be excessively long. This effect may be clearly seen from the two grid potential waveforms shown in Fig. 14. Waveform (a) is that of a blocking oscillator functioning normally, whereas (b) illustrates the effect of insufficient transformer damping, where the commencement of the linear rise of grid potential becomes somewhat delayed.

It will be noticed from (a) that very large negative pulses are possible at the grid, especially if a low repetition rate is desired, requiring larger grid inductances. These pulses are, however, of such small duration as not to affect grid-cathode insulation, and as far as the author is aware no cases of breakdown due to this cause have been experienced.

Turning now to the scanning period. From a consideration of Fig. 15, representing the potential across Cwith respect to time, an expression may be derived for the repetition rate of the sawtooth cycle in terms of the circuit time-constants and potentials. Commencing with the valve conducting, C is rapidly charged negatively to a potential V volts. With the valve biased beyond cutoff, the capacitor discharges through R towards the H.T. potential  $V_1$ . When the potential across C reaches the grid base of the valve  $V_3$ , C becomes charged by grid current and the cycle repeats.

The general formulæ for the charge or discharge of a capacitor C through a resistor R from a potential V, in a time t seconds is given as:

$$v = V (1 - e^{-t/OR}) \dots (3)$$

. . . . . . . . . . . (4)

and at time 
$$t_1$$
,  
 $-V = V_1 (1 - e^{t_1/CR})$   
 $\frac{V + V_1}{CR} = e^{-t_1/CR}$ 

 $V_1$ 

at time t<sub>a</sub>

from equations (4) and (5)



Fig. 15. The potential difference across C

Let 
$$t = t_2 - t_1$$
 = discharging period for C

then: 
$$t = CR \log_{0} \left( \frac{V + V_{1}}{V_{1} + V_{2}} \right)$$

giving total period of oscillation as:-

$$P = T + t$$

where 
$$T =$$
 period of charge for C,  
thus frequency of sawtooth oscillation:

$$F = 1/P = \frac{1}{T + CR \log_{\bullet}\left(\frac{V + V_1}{V_1 + V_2}\right)}.....(6)$$

The frequency of oscillation can thus be controlled by variation of R, C or  $V_1$ . This latter is convenient as a wide range of control can be exercised with a comparatively small value of potentiometer by returning the charging resistor R to a potentiometer across the H.T. supply.

The coupling transformer is by no means critical in design, unlike the radar pulse generator where pulse shape is an important criterion. In order to ensure a rapid charging period for the capacitor a small step-up ratio is usual from anode to grid windings. The total inductance required can be derived from equation (3), given the frequency desired (noting that only a single half-cycle during the charge period is desirable) stray capacitance and mutual inductance between windings. These latter two factors will require estimating from the physical con-

figuration of the completed transformer. High coupling factor is important to prevent oscillations due to the leak-age inductance from lengthening the retrace time. In In practice, the transformer is wound on a laminated iron core and the coupling factor exceeds K = 0.8 and is sufficiently large for this effect to be negligible.

One disadvantage of the simple blocking oscillator described above is that an inductance is included in series



Fig. 16. (a) Electron coupled blocking oscillator; (b) blocking oscillator with separate charging valve

with the charging circuit. This may be avoided by using a pentode, after the manner of the electron-coupled oscillator, where the blocking oscillator circuit is included between cathode,  $g_1$  and  $g_2$ , while a separate CR circuit is included in the anode circuit, (Fig. 16(a)).

Alternatively two valves may be used, one to act as the oscillator while the other, having its control grid strapped

#### **Electronic Peak-Reading Kilovoltmeter**

#### By M. Lorant

An instrument that rapidly and accurately measures the high voltages associated with medical X-ray equipment has recently been developed by the U.S. National Bureau of Standards. The unit is designed to indicate directly peak voltages between 30 and 125kV.

In principle, the kilovoltmeter operates as follows: A type 2D21 thyratron tube is used as the high voltage detector. The critical bias on the tube is about -3V at an anode potential of about 150V. If, for example, the bias voltage is 103V, the thyratron will fire when the voltage drop across two  $50k\Omega$  resistors between the bias cell and ground and the cathode and ground reaches 100V. These two resistors form part of a two-section voltage dividing network across the X-ray tube. Because the ratio between the sections of the voltage divider is nearly 1 000:1, the metering circuit will respond when the peak voltage across the X-ray tube reaches 100kV. Hence, any desired voltage may be measured by suitably adjusting the bias voltage.

The actual circuit of the Bureau's kilovoltmeter consists of eleven 2D21 thyratrons, each having an indicating neon light in its anode circuit. The grids of the thyratrons are biased in approximately one volt increments by potentiometers inserted into each grid circuit. Twenty series-connected resistors are connected to a 22-position selector switch so that a voltage adjustable from 30 to 125V in 5V steps may be added to the bias on each tube. These resistors and the grid potentiometers constitute a voltage dividing network across which 150V D.C. is applied. It is thus possible to pre-set the instrument to read the kilovoltage within any desired 10kV range, the reading being the sum of the setting on the coarse kV selector switch and the number indicated by the last activated neon lamp. A reversing switch between the voltage divider and the kV selector switch permits measurements of either the positive or negative peaks. A calibrating circuit is included for setting the bias voltages

A calibrating circuit is included for setting the bias voltages of the eleven thyratrons. When the "use-calibrate" switch is closed and the coarse kV selector switch is set at zero, any

to the oscillator grid, controls the charge and discharge of the capacitor in its anode circuit, (Fig. 16(b)).

In either case the capacitor is charged slowly from the H.T. supply and discharged rapidly through the valve impedance. It is thus necessary to maintain a low internal impedance only during the conductive period, the retarding effect of the transformer inductance not being included in the circuit.

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The basic circuit

desired voltage from 0 to 10 volts may be applied to the thyra-tron grids. This voltage is increased slowly and the bias poten-tiometers set until at each integral voltage the corresponding thyratron fires.

A time delay circuit, incorporating two thyratrons and two relays, is incorporated to prevent recording of the starting surge or of any surge occurring at the end of an exposure, includes two thyratrons and two relays.



## EUROPEAN TELEVISION and the Coronation

THE growth of television services in Europe, compared with what has taken place in North America, is at a very slow rate indeed.

There are in the United States alone some 150 television transmitters operating on a regular basis and many of the larger cities have more than one transmitter. In New York a viewer can, if he feels so disposed, sit in front of his receiver for nearly twenty-four hours of the day and take his choice of the thirteen transmitters which radiate from there. For important national events such as the recent Presidential elections of last year, vast audiences of many millions—there are over 23 million television receivers in use—can be assembled by the coast-to-coast relay network which now extends from New York and Florida across to Hollywood.

To this vast American television system has recently been added the Canadian network which opened last September with transmitters at Toronto and Montreal.

Outside the North American continent the only country which can claim comparable progress is Great Britain. A regular service—the first in the world—was actually operating as far back as 1936 from Alexandra Palace, but it was not until after the war that a scheme was prepared for national coverage which envisaged a chain of five high power and five medium power stations giving coverage to some 90 per cent of the population of the United Kingdom. The last of the high power stations was brought into use in the summer of last year, but work on the remaining five medium power stations has not yet been granted Government sanction.

America and Great Britain each have a unified television

standard and the limiting factor in the further growth is largely one of finance. In Great Britain matters from the technical point of view are simplified by the fact that there is a single network radiating a single programme for the whole country and consequently the simplest type of receiver is all that is at present required, although matters may be slightly more complicated when sponsored television starts in this country.

But when one turns to Europe, a very different picture presents itself. There are many small but densely populated countries bordering on one another, each with their separate languages and national interests, yet few are able to maintain à regular television service.

There was a pre-war television service operating on a 441 line system from Paris and this was augmented after the war by high definition 819 line transmitters at Paris and at Lille, but elsewhere in Europe television development has been entirely post-war and on a very small scale only.

The exchange of programmes had been put forward as a means of accelerating the growth of television in the various countries of Europe, but considerable technical difficulties lay ahead arising from the different television standards adopted. Indeed, direct exchange and simultaneous broadcast of programmes might have been ruled out entirely but for the development by the BBC Engineering Department of a convertor unit.

The relay from Calais to Britain in 1950 was, in reality, an extension of the O.B. system of the BBC in which no conversion was involved, and the first direct exchange of programmes between countries operating on different standards took place last year when programmes originat-





#### The television network for the Coronation broadcast

Place names on the coaxial cable links indicate the repeater stations. Place names on the radio links indicate relay voints. London-Wenvoe cable route is not shown

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ing in Paris were simultaneously radiated over the French and British networks. In this case a convertor unit was installed at Cassel in Northern France to convert from the French to the British standard.

This exchange showed that direct exchanges were quite possible from the technical point of view and in view of the world wide interest in the British Coronation it was only logical that an attempt should be made to relay the Coronation television programme to Europe and to provide new relay links where required.

To this end, therefore, representatives of the Institut National Belge de Radiodiffusion, The Nederlandse Televisie Stichting and The Nordwest-Deutscher Rundfunk agreed to work out plans in conjunction with the British Broadcasting Corporation and the Radiodiffusion-Télévision Française, who were responsible for the technical arrangements of the Franco-British week of television in July, 1952. A temporary relay network has now been set up across Belgium and Holland linking those countries and Germany with France and Britain, and full scale



Mobile temporary transmitters are being installed at Pontop Pike, County Durham, and at Glencairn, near Belfast, in time for the Coronation. The programme input equipment bays and one of the duplicate vision transmitters are shown above. (Marconi's Wireless Telegraph Co., Ltd.)

tests have been recently carried out during which the normal BBC transmissions were successfully relayed to the countries concerned. In all some 27 repeater links are involved along a route of 900 miles across the Continent.

Probably the most convenient manner of describing the actual relaying of the Coronation across Europe is to start with the cameras themselves and to trace the path of the television signals stage by stage.

On the Coronation Day some twenty-one cameras will be in use located at five main sites as follows :---

The Victoria Memorial opposite Buckingham Palace. Inside Hyde Park near Hyde Park Corner.

The Colonial Office overlooking the west front of Westminster Abbey.

The Embankment near Westminster Pier. Inside Westminster Abbey.

cameras, to-These gether with the various control and commentator points, will be con-nected to Broadcasting House by cable networks provided by the Post Office. From Broadcasting House the television signals will be trans-mitted by the five high power stations which are connected to Broadcasting House by cable and radio links as shown on the accompanying map. A further cable link connects the studios at Lime Grove, where film recordings will be made.

In addition to the main transmitters, low power transmitters have been erected at Glencairn (Belfast), Pontop (Newcastle) and Pike Trueleigh Hill (Brighton). These have been erected on a temporary basis only, but will remain in operation until the approval of the Government has been obtained for the erection of the power medium five stations which form the remainder of the BBC's plan for national coverage.

The Glencairn transmitter will be connected to the main BBC network by a tem-porary radio link pro-vided by the Post vided by the True-Office, while the True-Hill transmitter will receive its signals by direct pick-up from Alexandra Palace. The temporary transmitter at Pontop Pike is installed on the site allocated for the permanent medium power transmitter and consequently it will receive its signals direct from the permanent television link.

The television signal on its journey across Europe starts from the Senate House of the University of London. which is connected to Broadcasting House by cable. From Senate House the signal is relayed by an S.H.F. radio link to Wrotham in Kent and thence by a similar link at



235 ft. mast with 231 ft. singlestack super turnstile aerial (horizontal polarization) which Marconi's Wireless Telegraph Co.. Ltd., have erected at Glencairn Road, Belfast, in time for the Coronation

The scaffolding structure near Calais erected to mount the two receiving aerials for the Dover-Calais link and the transmitter for television signals to Cassel. The cross channel link was arranged with "space diversity" to ensure good reception



#### ELECTRONIC ENGINEERING

**JUNE 1953** 

Lenham to Swingate near Dover. (The aerial assembly on the roof of Senate House is shown in the photograph at the top of page 244).

From Swingate the signal is relayed across the English Channel to Cap Blanc Nez, on the coast of France, where it is relayed to Cassel.

The whole of the equipment in this circuit from London to Cassel has been provided by Standard Telephones and Cables Ltd., under contract from the BBC and RTF.

The "Standard" portable S.H.F. link comprises lightweight equipment, forming a complete transmitter and receiver operating in selected bands in the frequency range 3 600-4750Mc/s. The transmitter has a velocity modulated coaxial line valve delivering a power of 250mW into a four foot diameter paraboloid. A similar valve is used as the local oscillator in the receiver. The limiting of the power delivered to the aerial makes possible the use of a directly modulated coaxial line oscillator feeding direct to the aerial so that all the apparatus forms an integral part of the aerial and paraboloid assembly.

The output of a receiver unit can be connected directly to the input of a transmitter unit and a limited number of links may be so connected in tandem.

The system is frequency modulated with a deviation of 4Mc/s although a deviation of 8Mc/s can be obtained, with some loss of linearity, where difficult transmission conditions exist.



The repeater station near Calais which occupies a German dugout in the Cap Gris Nez area. The photograph shows the two sets of portable units.

From Cassel the signal, still at the British 405 line standard, is relayed to Paris by further radio links along the route of the Paris-London relay of last year. At Paris convertor equipment is installed to convert to the French 441 and 819 line standards of the two Paris television transmitters, the 819 line transmitter at Lille being connected to Paris by a further radio link.

There is at the moment no regular television service in Belgium although some parts of the country are in the service areas of the French transmitter at Lille and of the Dutch transmitters at Lopik and Eindhoven. It is . hoped that there may be a limited closed loop system in Brussels in time for the Coronation from a direct pick-up at Flobecq from the Lille transmitter.

The INR have undertaken to relay the television signal across Belgium from the relay point at Lille to the Dutch frontier at Breda, through intermediate relay points at Flobecq, Brussels and Antwerp with centimetric radio link equipment, supplied by Philips.

There is a regular television service in Holland operated by the NTS with a permanent transmitter at Lopik with studios in Bossum. This transmitter located in the centre



A view of the receiving terminal equipment at Cassel. This comprises a main and standby set of receiver control circuits. On the right may be seen the special large screen monitoring equipment provided by Kolster Brandes Ltd.

of Holland covers about 40 per cent of the population and is connected by a radio link to a low power experimental transmitter at Eindhoven.

Breda, on the Dutch frontier, serves as a convertor point where the British 405 line television signal is converted into the 625 European standard by a convertor of Philips manufacture. From Breda a radio link carries the signal, now at 625 lines, to Lopik and further radio links at Eindhoven, Helenaveen, and Suchteln carry the 625 line signal to Cologne, where it feeds the NWDR television network for the television transmitters at Cologne, Langenburg, Hanover, Hamburg, Berlin and Frankfurt.

This temporary route, of course, only covers those countries where television services are operating. As the remaining European countries develop their systems the network will be extended. In Denmark, for example, there is an experimental station broadcasting intermittently at Copenhagen. It is not certain at this stage whether Copenhagen can be connected by link to Hamburg. Should this not be so, film recordings of the Coronation made at Hamburg will probably be flown to Copenhagen for rebroadcasting. Of the remaining European countries, Sweden does not expect to start from Stockholm until the autumn of this year and Switzerland has a small experimental station near Zurich. There are three television transmitters operating in Italy and two more are due for completion this year. Spain has two small experimental stations, but no development plans have been announced. Much of the television relay network just described is on a temporary basis only and will be removed as soon as the Coronation ceremonies are over.

Interior view of the Philips mobile convertor stationed at the foot of the Grote Kerk in the Breda market square.



### A Television Waveform Display Apparatus

(Part 2)

By G. N. Patchett\*, Ph.D., B.Sc., A.M.I.E.E., M.I.R.E., A.M.Brit., I.R.E.

#### Performance

Having described the apparatus, some results will now be shown. The waveform (negative picture signal) of two lines on a BBC transmission of a moving picture is shown in Fig. 8(a). This should be compared with Fig. 1(a), a typical result on a normal oscillograph. Fig. 8(b) shows the frame synchronizing period (odd frame, positive picture signal) as received in the centre of Bradford from Sutton Coldfield (approximately 100 miles away). The line pulses, the half line, and eight frame synchronizing pulses can be distinguished. Although the interference pulses extend off the oscillogram a recognizable picture can be obtained under these conditions. Fig. 8(c) (negative picture signal) shows part of the frame period of early test signals (before official tests started) from Holme Moss. The "picture" was a white cross and it will be seen that only four half line synchronizing pulses were transmitted instead of the normal eight.

The waveform of approximately two lines of test-card C is shown in Fig. 9(a), while in Fig. 9(b) is shown the received picture on a television set with the corresponding two lines brightened by the pulse from the waveform display apparatus. An index to the various portions of the waveform is also given so that the waveform can be related to the actual picture. The actual picture is taken from a commercial set, but the connexion of the apparatus to the cathode of the picture tube upsets the high frequency response of the video stage and causes some lack of definition on the picture.

Fig. 10 shows in a similar way the frame synchronizing period on even frames, one set of flyback lines now being brightened by the brightening pulse. The oscillogram clearly shows the eight half line pulses followed by part of the blanking period. In. Fig. 11 is shown the odd frame synchronizing period, but at a reduced scan speed, a number of lines being shown before and after the period. The corresponding lines are shown brightened in Fig. 11(b).

Fig. 12 shows the results obtained on two frame synchronizing separator circuits. (a) Shows the effect of integrating the frame synchronizing pulses, this oscillogram being taken for the case of even frames. This oscillogram clearly shows the way the frame synchronizing pulses are lifted above the small line pulses and also shows the shape of the pulse at the start and, what may be equally important, at the end. (b) Shows the effect of passing the odd frame synchronizing pulses through a differentiating circuit. The original pulses were negative going, the pulses rising during the frame pulse period due to the change of D.C. component. This oscillogram shows the pulses clearly which is not possible on a normal oscilloscope.

As well as being able to view the video signal as it passes through the video amplifier and synchronizing separator circuits, the flyback of the time-base can be seen in detail. Although the apparatus was intended for television work it may also be used to view a small portion of any other waveform which contains small detail. Fig. 13(a) and (b) shows a small portion (approximately 1 per cent) of the approximately sinusoidal current taken by a high pressure mercury vapour discharge lamp. It will be noticed that over this portion small damped oscillations occur, presumably due to restriking of the arc in the lamp. The frequency of these oscillations is approximately 300kc/s and they only occur on this small portion of the



Fig. 8. Oscillograms taken with waveform display apparatus of television waveforms as received from the BBC on a receiver
(a) Typical waveform with moving picture showing just over two lines. Negative picture signal. Holme Moss transmitter.
(b) Frame synchronizing period (odd frame) received from Sution Coldfield in the centre of Bradford. Positive picture signal.

(c) Frame synchronizing period transmitted by Holme Moss during unofficial test period. Negative picture signal.

<sup>\*</sup> Bradford Technical College













#### Fig. 9 (top left). (a) Waveform on display apparatus of portion of test card C. Negative picture signal. (b) Picture on television receiver with corresponding lines increased in brilliance

brilliance A. Line synchronizing pulse. B. Back porch. C. Vertical white line. D. Black portion on left of circle. E. White line on black portion D. F. Left and right hand 2Mc/s bars. G. Grey centre square. H. White circle. I. Black circle. J. White portion on left of circle. K. Black line on white portion J. L. Vertical white line. M. Castellated edge of picture. N. Front porth.

Fig. 10 (bottom left). (a) Waveform on display apparatus of frame synchronizing period (even frames) during transmission of test card C. Negative picture signal. (b) Picture on television receiver with corresponding lines (now flyback lines) increased in brilliance

Fig. 11 (top right). (a) Waveform on display apparatus of test card C. Negavive picture signal. (b) Picture on television receiver with corresponding lines increased in brilliance A. Castellated edge of picture at bottom. B. 8 frame synchronizing pulses. C. Blanking period. D. Castellated edge of picture at top. E. Lines corresponding to the vertical lines on the picture and diagonal lines in the corner.

Fig. 14 (bottom right). Oscillogram taken with modified waveform display apparatus showing odd and even frame pulses after integration



Fig. 12. Waveforms on display apparatus of frame synchronizing period:----After (a) integration (b) differentiation



Fig. 13. Oscillograms taken with waveform display apparatus showing oscillations produced on a portion of the cycle of current taken by a high pressure mercury discharge lamp



Fig. 15. Oscillograms taken with modified waveform display apparatus on blocking oscillator time-base

- (a) Synchronizing pulse (inverted and
- applied to anode Anode waveform. Grid waveform. anode).

Output waveform. oscillograms are to the same hori-zontal scale. λŰ

cycle of current and are quite invisible on a normal oscilloscope.

A modification of this apparatus which proved invaluable has been used by the author on some research on interlacing<sup>7</sup>. For this purpose it was necessary to operate the time-base at 50c/s instead of 25c/s, so that odd and even traces are superimposed, it then being possible to see any difference on odd and even frames. The actual waveform display apparatus could have been used by cutting out the frequency divider and delay circuit, but the



ime of four lines

display apparatus was required for other purposes. A timebase was constructed exactly similar to the one shown in Fig. 6 and the output from this was fed to the X-deflecting plates of a cathode-ray oscilloscope and the brightening pulse was fed to the grid. The research was performed using a pattern generator as signal source. It was necessary to trigger the time-base exactly at the start of the frame pulses. This was done by deriving simple pulses (four lines long) from the pattern generator which were at frame frequency and identical on odd and even frames. This pulse was differentiated, reversed in phase and used to trigger the time-base  $(g_s, of valve V_{12}, of Fig. 6)$ . Fig. 14 shows a result obtained in this way, which consists of an oscillogram of the integrated frame pulses. It will be seen that there are two distinct pulses, one for odd frames and one for even frames, the start and end of the pulses being different in the two cases.

Fig. 15 shows some oscillograms taken in this way on a

blocking oscillator time-base circuit. The circuit was synchronized on the anode by the pulse shown in (a) (after phase reversal) which is four lines long. (b) Shows the voltage on the anode, where it will be noted that a damped oscillation is produced by the end of the synchronizing

Fig. 16. Oscillogram taken with modified waveform display apparatus on blocking oscillator time-base show-ing two different scans on odd and even frames, although the flyback is the same in both cases



pulse. The grid voltage is shown at (c) where again the damped oscillation can be seen and also the period during which the grid is driven positive. The output voltage is shown at (d) which shows the short flyback corresponding to about 2 lines. In Fig. 16 is shown an oscillogram on a similar time-base circuit, except that the time-base is synchronized (on the anode) by integrated frame pulses after passing through a limiter. It will now be seen that the flyback in the two cases is identical, but that two separate scans are produced, corresponding to odd and even frames. This is caused by the differences in the end of the integrated pulses (see Fig. 14 and Ref. 7). This difference in the two scans causes failure of correct interlace.

#### Conclusion

The waveform display apparatus has proved to be a very valuable addition to the television equipment and enables a student to see exactly what happens as a signal passes through the various stages of a set. It is also invaluable for fault finding on "sticky" faults connected with the synchronizing circuits. As well as being so valuable in servicing classes it has proved extremely useful in research on synchronizing separators and in its modified form enabled conclusive results to be obtained in connexion with the problem of interlacing.

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### An Electronic Current Stabilizer for a Low Resistance Solenoid

#### By M. S. Wills\*, B.Sc., M.A., A.Inst.P.

The 20 $\Omega$  solenoid is supplied from a rectifier set with direct currents up to 5A. Voltage variations across a standard resistance in series with the load are amplified and fed back to a control unit which adjusts the A.C. input voltage to the rectifier. The design and performance of the equipment are described.

THE stabilizer described in this article was constructed to meet the need for stabilization of a small water-cooled solenoid supplied by a rectifier set. The resistance of the solenoid was too low and its current consumption too high for a circuit of the world time containing a carine too high for a circuit of the usual type, containing a series control valve, to be suitable; such a circuit has been des-cribed for example by Graham, Harkness, and Thode<sup>1</sup>. The difficulty, which arises from the order of magnitude of anode resistance of the control valve, was overcome by using a control unit circuit described by Patchett<sup>2</sup> the action of which may be looked upon as a transformation

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of the valve anode resistance to a lower and more convenient value. The degree of stability needed was not exceptionally high, and a maximum current variation of  $\frac{1}{2}$  per cent could be tolerated. Higher stability could however be obtained if required by using an amplifier of higher gain.

#### Design

A block diagram of the equipment is shown in Fig. 1. Current from the 3-phase 356V 50c/s mains passes through the control unit into the rectifier unit. Direct current of 5A maximum at 170V from the rectifier flows
through the load in series with resistors  $R_{14}$  and  $R_{15}$  each of which consists of several rheostats in series. The resistor  $R_{14}$  has very heavy power rating in relation to the current it has to carry so that change of resistance due to heating is negligible. The voltage across  $R_{14}$ , which is thus proportional to the load current, is amplified and fed back to the control unit so as to compensate for changes of load current by adjustment of the input voltage to the rectifier unit.







Fig. 2. The control unit



The circuit of the control unit is shown in Fig. 2, and is repeated in each phase line, except for the part enclosed by dotted lines. The transformer connexions are arranged so that anode and screen are excited in phase so as to obtain a triode type characteristic in the way explained by Patchett<sup>2</sup>. The resistance connected to the secondary of  $T_2$  is  $R_2$  with  $V_1$  in parallel and the effective resistance of  $V_1$  is altered by change of the grid potential derived from the D.c. amplifier. The range of variation of grid potential is 0V to -50V. The secondary voltage of  $T_2$  depends on the load and on the grid potential of  $V_1$ . On full load and with  $V_1$  cut off this voltage rises to a maximum of about 600V. The choke  $L_1$  of about

10 turns is needed to counter parasitic oscillations of very high frequency.  $R_s$  is a grid stopper and the switch  $S_s$  is used to cut out the stabilizing action when the load current is being changed. A peculiarity of the circuit is that the controlling effect becomes less as the load current is reduced. It is thus necessary to have a dummy load taking about 2A ( $R_{14}$  in Fig. 1) to be switched in when the true load current is small.

Fig. 3 shows the single stage D.C. amplifier. The voltage developed across  $R_{14}$  which ranges from about 30V to 120V depending on the load current and the value of  $R_{14}$  is applied to the small wire wound potentiometer  $R_4$ and the latter is set to deliver a standard voltage, actually and the latter is set to deliver a standard voltage, actually about 24V, to the grid of  $V_3$  and this is opposed by the reference voltage of the dry battery  $B_2$ .  $R_4$  is to prevent the grid of  $V_3$  going appreciably positive should  $R_4$  be turned up too high. The voltage gain of  $V_3$  is 170. The output voltage of  $V_2$  is applied to the cathode-follower valve-voltmeter  $V_3$ . M is graduated to read directly the cutout voltage appearing at XY. The acorn diode  $V_4$  is output voltage appearing at XY. The acorn diode  $V_4$  is a safeguard to prevent the grid of  $V_1$  becoming positive.  $C_1$  is to stop parasitic oscillations. The H.T. supply for this unit is provided by a normal power pack fed from a constant voltage transformer and stabilized by V<sub>s</sub>, V<sub>s</sub> and V<sub>7</sub>.

#### Performance

To use the equipment, after the supplies have been switched on and with  $S_2$  switched to  $B_1$ , the load current is adjusted to the desired value by means of  $R_{14}$  and  $R_{18}$ .  $R_{14}$  and  $R_{15}$  each consist of several rheostats which are mounted and wired in such a way that, when they are cut out one by one in succession, the voltage across  $R_{14}$ is always greater than the standard voltage needed for stabilization.  $R_{16}$  is cut out by means of  $S_1$  if the load current exceeds about 2A.  $R_4$  is then adjusted until the meter M reads about 24V and  $S_2$  is thrown over to stabilize the current.

There are three causes of load current variations, namely, drift, variations of the supply voltage, and changes of load resistance due to heating and cooling. The latter effect is particularly important with a load such as a solenoid or electromagnet since the windings are of copper and are often run up to a fairly high temperature to obtain high performance. It is desirable to "swamp" this effect as far as possible by series resistors of low temperature coefficient. The drift of the equipment has been tested by using a load in which change of resistance due to heating was negligible, and setting the input voltage before each observation to a standard value. The drift of the current under these conditions is about 1 part in 1000 in 3 hours. The effect of change of input voltage varies according to the current and is least on full load. With a line to neutral voltage of 210V the effect of a change of  $\pm 5V$  is 0.21 per cent when the load current is 4.7A and 0.42 per cent when the load current is 0.17A. The voltage range over which the stabilizer acts is 23V in the former instance and 14V in the latter. The effect of increasing the load resistance is greatest with the largest current and less for smaller currents owing to the swamping effect of the rheostats. Thus at 4.5A the effect of an increase of  $3\Omega$  is to reduce the current by 0.6 per cent. Since the load resistance is about  $20\Omega$  copper this corresponds to a rise of temperature of about 40°C.

#### Acknowledgments

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### A Portable High Speed Stroboscope

### By A. E. Ferguson<sup>\*</sup>, M.E.E., A.M.I.E. (Aust.)

A light source stroboscope is described using an xenon filled tube giving a substantially white light output lasting for approximately  $30\mu$ sec. The flashing rate can be varied from 20c/s up to 500c/s. The whole equipment is mains operated and can be made readily portable.

THE electrical light source stroboscope, first developed by Edgerton and others<sup>1</sup> in America during the early 1930's, has proved a most versatile instrument in the hands of scientists and engineers. Although the gas discharge flash tube does not produce the short duration high intensity light output obtained from high voltage sparks a technique which has been used for photographing projectiles in flight since the close of the last century<sup>3</sup>—it is considerably more convenient and can be triggered by a variable frequency oscillator for stroboscopic work. The flash obtained from a modern tube is sufficiently brilliant to allow stroboscopic observations to be made under conditions of almost normal illumination and has a duration sufficiently short for some ballistic purposes<sup>3</sup>.

The development of flash tubes received a considerable impetus during the 1939-45 period by the demand for a high intensity light source for night photography. The British work in this period is well summarized by Aldington and Meadowcroft<sup>4</sup>.

A recently developed stroboscope<sup>5</sup> is capable of giving up to 4 000 flashes per second for high speed photography, but is not sufficiently portable for many requirements.

but is not sufficiently portable for many requirements. The instrument to be described uses one of the flash tubes which has been produced as a result of these war-

time developments. In association with auxiliary equipment, it can be used as a light source stroboscope giving up to 500 flashes per second and is readily portable.

### General Description of Stroboscope

A block diagram is shown in Fig. 1. A capacitor, charged to about 3kV, is discharged through the gasfilled flash tube upon the application of a short duration triggering pulse. The peak voltage of this triggering pulse is also of the order of 3kV.

A multivibrator forms a convenient variable frequency oscillator which determines the trigger pulse repetition rate. In this design the repetition rate may be varied from 20c/s to 500c/s. To achieve a reasonably open scale two 5 to 1 ranges have been used. A high and a low

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voltage supply allow the whole equipment to be mains operated.

A complete circuit diagram is given in Fig. 2.

### **Flash Tube Circuit**

The flash tube used is a Mullard type LSD8 which is xenon filled. The discharge takes place between two electrodes sealed into the ends of a glass helix. Under normal conditions where the discharge current is large the colour of the light emitted is approximately white. The trigger electrode is a helix of wire wound coaxially with and outside the glass helix.

The tube is designed primarily for stroboscopic applica-



Fig. 1. Block disgram of the stroboscope





tions. The mean dissipation should not exceed 30 watts.

(With forced air cooling, this rating may be increased). As already stated, a "storage" capacitance is discharged through the flash tube. For stroboscopic work, care must be taken that the wattage loading of the tube is not exceeded at high rates of flash.

If the storage capacitance is always recharged between flashes to the same potential, then the intensity of the flashes remains constant over the whole range of flashing speeds, but the wattage loading of the flash tube decreases in direct proportion to the frequency. At low flashing speeds it is clear that the maximum illumination capa-bilities of the tube cannot be realized without increasing the size of the storage capacitance. By careful design of the time-constant of the recharging circuit it is possible to maintain the wattage loading of the flash tube reasonably constant over a wide range of speeds without changing any components of the circuit. This means that the energy per flash increases as the flashing rate drops.

It is normal practice for this storage capacitance to be charged through a series resistance from a much larger capacitance shunting the output of the rectifier. This is used in the present design.



An analysis of the recharging of the storage capacitance and the loading of the flash tube is given in Appendix 1 where it is shown that the wattage dissipated in the flash tube is:

$$\frac{E^2C}{2T} (1 - e^{-T/RC})^2$$
 watts.

To determine the optimum value of the time-constant to use, the function

$$1/T (1 - e^{-T/RC})^2$$

may be plotted for a range of values of T/RC. This is shown in Fig. 3 where, for convenience, RC = 1.

It is clear from this graph that in the range T = 0.5RCto T = 3.0RC the function

$$1/T (1 - e^{-T/RC})^2$$

varies from 0.3 to just over 0.4. That is, for a 6 to 1 change in frequency of the flashing rate, the wattage loading may be kept within 25 per cent of the maximum wattage at any point of the range.

The larger the ratio T/RC, the greater the voltage build up across the flash tube between triggering pulses. This is shown in Fig. 4. If the trigger interval is of the order of 0.5RC the voltage build up will be only 40 per cent of the maximum and some difficulty may be experienced in obtaining satisfactory operation at the high speed end of the range.

During the experimental development of the circuit it was found that when T/RC = 0.5 the tube very occasionally missed firing. It was felt that insufficient recharg-

ing of the capacitance was responsible for this erratic behaviour. For this reason a design minimum for the ratio T/RC of 0.7 was adopted. This corresponds to the storage capacitance recharging to half the maximum voltage. A 5/1 range of flashing speeds is then obtained with the wattage loading within 33 per cent of the maximum.

Using 
$$T/RC = 0.7$$

and 
$$T_{\min} = 2$$
 milliseconds (i.e. 500c/s)

$$E = 3kV$$

Wattage Loading = 30 watts (maximum) the constants of the charging circuit can be evaluated.

They are:

 $C = 0.046 \mu F$ 

 $\boldsymbol{R}=6\cdot 2\times 10^{4}\Omega$ 

The lower range from 100 flashes per second to 20 flashes per second uses a 0.25µF capacitor which is connected to the range change switch on the multivibrator. The maximum energy per flash with this capacitance is well within the rating of the flash tube. For stroboscopic



work the light output obtained per flash from the discharge is ample for most purposes.

### **Trigger Pulse Amplifier**

The oscillator controlling the repetition rate is a con-ventional multivibrator operated from a stabilized voltage supply. The frequency is varied by adjusting the positive bias and switching the coupling capacitance. The charac-teristic square wave output of this circuit is "sharpened" in a 6J6 squaring amplifier and coupled to the grid of the EL38 trigger amplifier by a 40 µsec time-constant circuit. The EL38 is a high transconductance output pentode capable of handling high peak anode voltages. The anode circuit of this trigger amplifier is operated from the high voltage rectifier output through a  $0.5M\Omega$  load resistor. The quiescent grid bias is below the cut-off value, hence a short duration negative pulse of approximately 3kV peak is developed in the anode circuit each cycle of the multivibrator. This is directly coupled to the flash tube trigger electrode.

Provision has also been made for synchronizing the multivibrator to the 50c/s mains supply.

The high tension rectifier is a half-wave mercury vapour hot cathode type with a shunt capacitance filter for smoothing. A 30sec time delay in the primary of the trans-former for this rectifier is included to allow the recommended warming up period.

#### Performance

The wattage loading of the tube has been measured at

various flash speeds and the experimental results are shown in Fig. 5. These are in reasonably good agreement with the results of the analysis given in Appendix 1.

The duration of the flashes has been measured using a multiplier photo-cell and an oscillograph. The photo-cell output was connected direct to the Y deflexion plate and





Fig. 6. Typical traces on screen of C.R.T.

the time-base triggered from the pulse at the grid of the trigger amplifier. Typical traces obtained direct from the screen of the cathode-ray tube are shown in Fig. 6. It is seen that the duration of the flash is of the order

of  $35\mu$ sec when the repetition rate is 20c/s. This corresponds to an energy per flash of the order of 1.0 joules. At the maximum repetition rate of 500c/s, corresponding to an energy of about 0.04 joules per flash, the flash duration is slightly less.

Acknowledgment

The author wishes to thank Professor C. E. Moorhouse for the facilities offered in his department for the development and construction of the instrument described.

#### APPENDIX

The equivalent circuit of the flash tube and its associated network is shown in Fig. 7. In this representation the switch S can be assumed to close instantaneously at each trigger pulse, so discharging the capacitor C. In the interval between flashes (when the flash tube is nonconducting) the switch must be open and the capacitor Cis then recharging from the high tension supply E through the resistor  $R_{.}^{-}$ 



Fig. 7. Equivalent circuit of flash tube

The voltage across the capacitor increases exponentially during the recharging interval, given by

$$e_{\rm o} = E(1 - e^{-t/BC})$$

where  $e_0$  = instantaneous voltage across the capacitor t = time measured from the instant the switch opens

E = D.C. rectifier output voltage

If the interval between flashes is T seconds, then

Energy per flash = 
$$\frac{CE^2}{2} (1 - e^{-T/BO})^3$$
 joules

Wattage loading = Energy flash x number of flashes per second

$$= \frac{CE^2}{2T} (1 - e^{-T/BO})^2$$
 watts

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### **Voltage Stabilization with Series Valve Control**

By R. D. Trigg\*, A.M.Brit., I.R.E.

It is intended to show in the following circuit analysis the criteria of voltage stability achieved by the circuit shown in Fig. 1.

FIG. 1 represents the conventional voltage stabilization metwork where mains supply fluctuations up to  $\pm 15$ per cent have negligible effect on the load voltage, when the load takes a large current.

Terminals A and B are normally connected to the smoothing filter and power supply system, and terminals C and D to circuits taking a relatively large current and requiring a stabilized high-tension supply.

When the mains voltage decreases, the corresponding reduction in  $V_{do}$  is communicated via  $V_1$  to the potential at c; and in turn, to the control grid of  $V_2$ . This results in a

rise in potential at the control grid of V<sub>1</sub>, and the voltagedrop across  $R_1$  is reduced.

Since the cathode potential  $v_{do}$  of  $V_1$  floats on the grid potential  $V_{a2}$  of this value, a reduction in  $v_{dc}$  is countered by the increased potential at  $V_1$  control grid. In order for this to occur, however,  $V_{a2}$  must change by a finite amount in excess of  $v_{do}$ . The degree to which the difference  $V_{a_2} - v_{g_1} = v_{d_0}$ , is made small will therefore mainly depend on  $V_2$  and its associated network.

It will be seen that a rise in mains voltage produces a similar result, the incremental change in potentials then having opposite sign.

<sup>\*</sup> Redifon Ltd.

Unless the potential at E is maintained constant with mains variation, cathode feedback will occur in the V<sub>2</sub> grid circuit, and the incremental change in  $V_{a_2}$  is reduced. This is avoided by employing the voltage stabilizer N, as shown.

For optimum results  $V_1$  should have a high amplification factor rather than a steep slope, and an anode dissipation which permits it to pass the full load current (i.e. two or more valves should operate in parallel to meet this requirement).  $V_2$  should possess a high mutual conductance.

### **Equivalent Circuit (Triode Operation)**

If the potential at E in Fig. 1 remains constant over the range of operation, the incremental change in potential at c with respect to D, i.e.  $\delta v_{do}$ , will also be the same as that with respect to E. Thus if only incremental variations in voltage and current are considered in the equivalent circuit, the stabilizer N may be omitted.

It is to be observed, however, that the stabilizer voltagedrop modifies the performance of the equivalent circuit, in so far as it decides the ratio  $v_{g_3}/v_{d_0}$ .



The current in resistors  $R_a$  and  $R_b$  in Fig. 1, which are chosen to give the correct bias to  $V_2$ , can in practice be made negligible in comparison with the load current. The quiescent potential at  $V_2$  grid may then be represented by  $i_1R_3$  where  $R_2 + R_3 = R_K = load$  resistance (see Fig. 2) and:

$$\frac{R_{a}}{R_{a}+R_{b}} = \frac{R_{2}}{R_{2}+R_{3}} \qquad (1)$$

$$R_{\underline{x}} = v_{do}/i_{1}$$

Thus from the figure :

$$\begin{aligned} \delta V_{do} &= \delta V_{a1} + \delta i_1 R_{\mathbb{K}} \qquad (2) \\ \delta V_{a1} &= \delta i_1 r_{a1} - \mu_1 \delta v_{g1} \qquad (3) \\ \delta v_{g1} &= \delta V_{a2} - \delta i_1 R_{\mathbb{K}} \qquad (4) \\ \delta V_{a2} &= \delta i_2 r_{a2} - \mu_2 \delta v_{g2} \qquad (5) \end{aligned}$$

Substituting Equations (3) to (6) in (2):

$$\delta V_{do} = \delta i_1 \left\{ r_{a_1} + R_{\mathbb{K}}(\mu_1 + 1) + \mu_1 \mu_2 R_s \right\} - \delta i_2 r_{a_2} \mu_1 \dots (7)$$
Also from Fig. 2:

$$\delta V_{d0} = \delta V_{a2} + \delta i_2 R_1 \dots (8)$$
  
and substituting Equations (5) and (6) in (8):

 $\delta V_{do} = -\delta i_1 \mu_2 R_3 + \delta i_2 (r_{a_2} + R_1) \dots (9)$ Solving for  $\delta i_1$ , from Equations (7) and (9):

$$\frac{\delta V_{dc}}{\left\{ \overline{r_{a_1} + R_{\mathbb{X}}(\mu_1 + 1) + \mu_1 \mu_2 R_a} \right\}, -\mu_1 r_{a_2}} = -\mu_2 R_a$$

$$\delta V_{de} \left\{ \frac{r_{a_2}(\mu_1 + 1) + R_1}{r_{a_1} + R_{\mathbb{K}}(\mu_1 + 1)_2(r_{a_2} + R_1) + \mu_1 \mu_2 R_1 R_2} \right\}. (10)$$

whence  $\delta i_1 R_{\mathbf{K}} = \delta v_{dc}$ 

$$= \delta V_{do} R_{K} \left\{ \frac{r_{a_{2}}(\mu_{1}+1) + R_{1}}{[r_{a_{1}} + R_{K}(\mu_{1}+1)](r_{a_{2}} + R_{1}) + \mu_{1}\mu_{2}R_{1}R_{3}} \right\}$$
(11)

and the stabilization ratio:

$$S = \frac{\delta V_{do}}{\delta v_{do}} = \frac{\left\{ r_{a_1} + R_{\mathbb{K}}(\mu_1 + 1) \right\} (r_{a_2} + R_1) + \mu_1 \mu_2 R_1 R_s}{R_{\mathbb{K}} \left\{ r_{a_2}(\mu_1 + 1) + R_1 \right\}}$$
(12)

..... (12)

since  $\mu_1 \ge 1$ , to a first approximation Equation (12) can be expressed as:

$$S = \frac{(r_{a_1} + R_{\underline{w}}\mu_1)(r_{a_2} + R_1)}{R_{\underline{w}}(r_{a_2}\mu_1 + R_1)} + \frac{\mu_1 R_3}{R_2 + R_3} \left(\frac{\mu_2 R_1}{r_{a_2}\mu_1 + R_1}\right).$$
 (13)

For triode operation:

$$R_{\mathbb{K}}\mu_1 \gg r_{\mathbf{a}_1} \qquad (14)$$

$$r_{n_2}\mu_1 \gg R_1 \qquad (15)$$

and the left-hand fraction in Equation (13) reduces to:

$$1 + K_1 / T_{a2}$$

while examination of the right-hand fraction shows that this is a more or less conventional expression for the effective gain of  $V_2$ . Within the brackets it will be observed that  $V_1$  increases the internal resistance of  $V_2$ ,  $\mu_1$  times. That is,  $V_2$  tends to operate as a pentode, due to the reason that a large variation in potential at A in Fig. 1 produces only a small change in potential at c by virtue of the self-biasing action of  $V_1$ . This, in turn, produces only a small fluctuation in grid potential at  $V_2$ , and consequently the plate current variation is small. This closely resembles the action of a pentode, and it is to be no ed this occurs without change in the original mutual conductance of  $V_3$ .

The expression adjacent to the brackets in Equation (13) represents the coefficient of the input voltage to  $V_3$  and is proportional to  $\mu_1$ .

On the assumption given in Equation (15), Equation (13) can be simplified to:

$$S = 1 + R_1/r_{a_2} + g_{m_2}R_1\left(\frac{R_3}{R_3 + R_3}\right).....(16)$$

$$(g_{m_3} = \mu_2/r_{a_3})$$

thus showing that  $V_2$  should possess a steep slope. Normally, triodes have steeper slopes than pentodes for the same anode dissipation, and a double-triode, with each section in parallel, serves extremely well in this stage.

#### Modified Circuit of Fig. 1.

The occasion sometimes arises where the load current is in excess of the current which can be taken by the valve or valves employed for  $V_1$ ; but the excess is not so great that it merits the use of another valve in parallel. This can be met by the inclusion of a resistor connected between anode and cathode of  $V_1$ , which accommodates the excess current, provided the deleterious effect this has on the stabilization ratio, does not reduce the latter below that required.

This shunt resistor causes  $V_1$  to behave virtually like a valve having an amplification factor equal to:

and internal resistance

where  $\mu_1 = \text{original amplification factor of } V_1$  $r_{a_1} = \text{original internal resistance of } V_1$  $R_4 = \text{value of shunt resistor}$ 

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the mutual conductance remains unchanged.

Equations (13) to (16) will then not apply, and rearranging Equation (12) in the form:

$$S = \frac{(r_{a_2} + R_i) \left\{ R_{\rm K} + r_{a_1}'(1 + R_{\rm K}g_{m_1}) \right\} + \mu_1' \mu_2 R_1 R_3}{R_{\rm K} \left\{ r_{a_2}(\mu_1' + 1) + R_1 \right\}}$$
(19)

indicates the effect  $R_4$  has on the stabilization ratio.

A fully worked example will reveal more clearly the operation of this circuit than the foregoing equations.





Capacitance is shunted across  $R_{3}$  to increase hum voltage e'r, applied to  $V_{3}$  control grid.

The arrangement shown in Fig. 3 (which is now in operation), has the following circuit parameters:

 $\mu_1 = 20$   $r_{a_1} = 2.5 kQ \text{ (two valves in parallel).}$   $g_{m_1} = 8 mA/V$   $\mu_2 = 32$   $r_{a_2} = 7 k\Omega \text{ (two valves in parallel)}$   $R_{K} = 3.62 k\Omega = 300 V/83 mA$   $R_1 = 22 k\Omega$  $R_4 = 12 k\Omega$ 

 $V_2$  has a quiescent bias of -4V and a constant cathode potential of 110V; hence the grid potential of this valve is 106V. The equivalent resistance from grid to ground, referred to the load resistance (see Equation (1)) is therefore:

$$R_{3} = R_{K} \cdot v_{g_{2}} / v_{do}$$
  
= 3 620 × 106/300  
= 1 280Ω.

The equivalent amplification factor of  $V_1$  with  $12k\Omega$  in parallel is (see Equation 17).

$$\mu_{1}' = 20 \left( \frac{12 \times 10^{3}}{2.5 \times 10^{3} + 12 \times 10^{3}} \right) = 17.5$$

and its equivalent internal resistance:

$$r_{a1}' = 2.5 \times 10^3 \times 0.875 = 2.16 k\Omega$$

Inserting these values in Equation (19) gives S = 32.4 (it is of interest to note that with  $R_4$  removed, S = 33.3). When the mains supply causes  $V_{dc}$  to fall by 32.4 volts, then:

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also

$$\begin{split} \delta V_{a_2} &= -\delta i_2 R_1 + \delta V_{d_0} \\ \delta i_2 &= -\left(\frac{\delta V_{a_2} - \delta V_{c_0}}{R_1}\right) = -\left(\frac{0.79 + 32.4}{22 \times 10^3}\right) \\ &= -1.51 \text{ mA} \\ \delta i_2 r_{a_2} &= -1.51 \times 10^{-3} \times 7 \times 10^3 = -10.57 \text{V} \\ \delta i_2 R_1 &= -1.51 \times 10^{-3} \times 22 \times 10^3 = -33 \text{V} \\ \mu_2 \delta v_{g_2} &= -(\delta V_{a_2} - \delta i_2 r_{a_2}) = -(0.79 + 10.57) = -11.36 \text{V} \\ \delta v_{g_2} &= -11.36/32 = -0.36 \text{V} \end{split}$$

This last result can be checked from:

$$\delta v_{g_2} = \delta v_{dc} \cdot R_3 / R_K = -1.0 \times \frac{1280}{3620} = -0.36V$$

The quiescent values of voltage and current and their corresponding stationary values for  $\delta V_{do} = -32.4$ V are, from the foregoing:

$V_{\rm do} =$	= 450V;	$V_{\rm dc} + \delta v_{\rm dc} =$	450 - 32.4	=	417·6V
$v_{\rm dc} =$	= 300V;	$v_{dc} + \delta v_{dc} =$	300 - 1	=	299V
$i_1 =$	83mA;	$i_1 + \delta i_1 =$	83 - 0·276	=	82·72mA
$i_2 =$	7mA;	$i_2 + \delta i_2 =$	7 – 1.51	=	5•49mA
$v_{g_1} =$	-10V;	$v_{g_1} + \delta v_{g_1} =$	-10+1·79	=	-7·21V
$v_{g_2} =$	-4V;	$v_{g_2} + \delta v_{g_2} =$	-4 - 0.36	=	4·36V
$V_{a_2} =$	290V;	$V_{a_2} + \delta V_{a_2} =$	290 + 1.79	==	291 <b>·8V</b>
$V_{a_1} =$	150V;	$V_{a_1} + \delta V_{a_1} =$	150 - 31.4	==	118·6V

### Hum Reduction

Since the circuit of Fig. 1 minimizes fluctuation in load voltage, it will be apparent that it also considerably reduces ripple voltage in the load. If the capacitor  $C_1$  were removed, the reduction in hum would be in the same proportion as the reduction in load voltage variation; i.e. the ripple voltage E impressed at terminals A-B would be reduced S times in the load. This assumes that the stabilizer N in Fig. 1 maintains a ripple-free potential at V<sub>2</sub> cathode. As this is not quite true, the stabilizer is sometimes shunted as shown with a capacitor  $C_2$ , when a very low hum level is desired.

By connecting  $C_1$  in parallel with  $R_2$ , the ripple voltage dropped across  $R_3$  is increased. Reference to Equation (13) will show that this increases the ratio S to a new value  $S_r = E_r/e_r$ , where  $e_r$  is the ripple voltage in the load.

The ratio of the ripple voltage  $e'_r$  impressed on the control grid of  $V_2$ , to  $e_r$  is determined as follows. The impedance of  $R_2$  and Z in parallel (see Fig. 4) is:

$$\frac{R_2Z}{R_2 + Z}$$

hence the impedance across which  $e_r$  is impressed is:

$$Z_{\rm r} = \frac{R_{\rm s} + Z}{R_{\rm s} Z} + R_{\rm s} = \frac{R_{\rm s} R_{\rm s} + Z(R_{\rm s} + R_{\rm s})}{R_{\rm s} + Z} \dots (20)$$

putting

where

$$Z = jKR_2 = -jX_c$$

whence

$$e_{\rm r}'/e_{\rm r} = R_{\rm s}/Z_{\rm r} = \frac{R_{\rm s}(1+K^2)}{R_{\rm s}+K^2(R_{\rm s}+R_{\rm s})-iKR_{\rm s}}$$
.. (22)

Usually  $C_1$  is made sufficiently large, so that Equation (22) becomes effectively equal to unity. The ripple voltage ratio  $E_r/e_r$  then becomes equal to:

$$S_{\rm r} = \frac{(r_{\rm a_1} + R_{\rm K}\mu_1)(r_{\rm a_2} + R_{\rm \cdot})}{R_{\rm K}(r_{\rm a_2}\mu_1 + R_{\rm \cdot})} + \mu_1 \left(\frac{\mu_2 R_{\rm \cdot}}{r_{\rm a_2}\mu_1 + R_{\rm \cdot}}\right) (23)$$

which approximates closely to  $g_{m_2}R_1$ .

### Logic and the Circuit Designer

### By T. L. Craven\*, B.Sc., A.M.I.E.E.

The increasing use of algebraic logic, or Boolean algebra, in the design of switching circuits must, one feels, cause the engineer to ponder on the nature of logic itself, a subject studied by the academics long before it found applications in technology. Accordingly, the aim of this article is to introduce Boolean algebra, and its uses, by way of some of the notions of the fascinating field of pure logic.

OGIC concerns itself with the passage from evidence to conclusions, that is, with inferences.

Applied logic busies itself with the physical world about us. Is it true that all thunderstorms turn milk sour? What evidence do we require to make such a generalization? How do we test its truth? These are typical of the These are typical of the questions that applied logic seeks to answer.

Formal logic, which is the support of algebraic logic, is concerned with the nature or form of inferences, but is indifferent to their subject matter. The argument which Indifferent to their subject matter. The argument which runs "Some amplifiers are magnetic devices, therefore some magnetic devices are amplifiers" is to the formal logician no different from that which runs "Some X's are Y's, therefore some Y's are X's". He wishes to know nothing about amplifiers or magnetic devices. The argu-ment, the inference, is the thing. Is it valid? If so, what are the consequences of such validity? But first it is necessary to explain the word valid. How often when you have developed some argument.

How often, when you have developed some argument, starting from what are known in logic as the premises, and arriving at what is known as the *conclusion*, must you have said that your conclusion followed necessarily from your premises. By that you had meant, perhaps not entirely consciously, that the conclusion followed from the premises simply because of the nature or the form of the argument, and not in any way because of its subject matter. You further meant that, while you did not guarantee that your premises were true, you insisted that your hearers, if they should choose to accept them as true, must then perforce regard the conclusion as true. And such is the point of view adopted when an argument is spoken of as being valid.

Validity is not the same thing as truth. The following arguments are all of the same form, and it is a valid one:

All humans are mortals

All logicians are humans ... All logicians are mortals.

All olotoxes are popocols All quintopods are olotoxes

... All quintopods are popocols.

All Y's are Z's All Xs' are Y's

 $\therefore$  All X's are Z's.

All males are rascals

- All engineers are males -
- : All engineers are rascals.

The first argument makes true statements, the second speaks of imaginary things, the third is purely abstract, and the fourth, one hastens to say, makes false statements. All, however, are valid. Of each one can say that, if the premises are true, then so, necessarily, is the conclusion. Of such matters does formal logic treat, epitomizing

the nature of validity in the statements that, in a valid argument, if the premises be true, then the conclusion must be true, and if the conclusion be false, then something in the premises must have been false.

#### **Testing for Validity**

We have now seen what validity is, and the bearing it has on implications of truth and falsity in the constituent statements of arguments. It is not enough, however, that logic should study such topics, for there remains also the matter of deciding whether any particular argument is or is not valid.

Logic has many rules for making such a decision. One of them, to give an example, is that we must not in the conclusion speak of "all X's" if in the premises we have only been presented with statements about "some X's". To do so would be to widen unjustifiably the scope of the available evidence.

It is this rule which tells us that it is not valid to argue

from All men are bipeds (premise)

All bipeds are men. (conclusion)

For in the conclusion we speak of all bipeds. In the premise, however, we have not done so, for the premise merely states "All men are some (and only *perhaps* all) of the bipeds in the universe.

From "All men are bipeds" we may therefore only infer that "some bipeds are men".

It is probably felt, though, that these arguments could well have been dealt with by common-sense, and that logic is being tedious; it must be remembered, however, that logic is not attempting to compete with common-sense, but is merely exploring a field of study. And common-sense is not always a very efficient weapon. Can you say with certainty whether or not it is valid to argue

Some quintopods are not men from

Some men are not quintopods? to

Think it over before you read on!

The answer is that the argument is not a valid one. It is invalid because in the premise we only speak of some quintopods, whereas in the conclusion we speak of all quintopods. You will be able to appreciate that better if we make our "some men" in the conclusion rather more specific. Let us suppose that, when we say "some men", we have in mind Tom, Dick and Harry. Then our conclusion really says that *all* quintopods have the property of not being Tom, Dick or Harry.

This may help to convince you—the true statement "Some humans are not engineers" does not imply that "Some engineers are not humans"!

### **Relationships**—and Contradictions

We turn now to an intriguing branch of formal logic which has a direct counterpart in switching circuits. This is the study of contradictory statements.

Two statements are said to be in contradictory relationship to each other when they are such that they cannot

<sup>\*</sup> Automatic Telephone and Electric Co. Ltd.

both be true and cannot both be false, so that one must be true and the other false.

It is so easy to regard statements as contradictory when they are not so, that it will be better to consider first some relationships that are not contradictory though deceptively near it.

Consider Fig. 1. It will be readily seen that (A) and (E), though they cannot both be true, do not satisfy the definition of contradictories because they can both be false (if some, but not all, men are fools). They are said to be in contrary relationship to each other.

Nor are (I) and (O) contradictories for, while they cannot both be false, they can both be true. They are in sub-contrary relationship to each other.





The true relationship of contradiction holds between (A) and (O), and likewise between (E) and (I). (A) and (O) cannot both be true, nor can they both be false. One must be true, the other false.

Where two statements are contradictory, then a knowledge that either one of them is true or false enables one to infer validly that the other is, respectively, false or true. Contradictories, therefore, stand in a more fruitful relationship for the making of inferences than do either contraries or sub-contraries. It a statement is true, one can infer that its contrary is false, but if a statement is false, one can infer nothing about the state of its contrary; a complementary shortcoming applies with sub-contraries.

Contradiction begins to be really intriguing when one turns to the study of *compound* statements. Each of the statements we have considered so far has made a single assertion, but a compound statement makes two or more assertions in combination. The two types of compound statements that are of greatest interest to us are exemplified thus :

(a) It was wet AND it was windy (which is called an "AND" statement),

and (b) EITHER he is hungry OR he is greedy (which is called an "EITHER/OR" statement).

What do you think is the contradictory of "It was wet and it was windy?" Would you hazard a guess that it is "It was not wet and it was not windy?" But those two statements are not contradictories, for it is possible for them both to be false (if it was wet but not windy, or if it was not wet though windy!).

So we shall have to look further afield and turn to a theorem propounded by Augustus de Morgan (English mathematician and logician, 1806-1871) for guidance.

De Morgan's theorem tells us that the contradictory of an "AND" statement is formed by contradicting its two halves and recombining them as an "EITHER/OR" statement—and vice versa. The respective contradictories of our two respective contradictories of our two original examples (a) and (b) are, therefore

(a) Either it was not wet or it was not windy, and  $(\beta)$  He is not hungry and he is not greedy.

On consideration it will be seen that

if	(a)	is.	true,	(a)	must	be	false,
if	(a)	is	false,	(a)	must	be	true,
if	<b>(</b> <sub>0</sub> )	is	tme	<b>(a)</b>	must	he	false

if (a) is false. (a) must be true,

and that similar implications hold between (b) and ( $\beta$ ).

The circuit (see Fig. 2) which is completed by the closure of switches A and B is broken by the opening of either A or B. That which is completed by the closure of either A or B requires, for its breakage, the opening of both A and B. Do you see, in each case, the likeness to De Morgan's theorem ?

### Algebraic Logic and Switching Circuits

Algebraic logic was not invented with any technological application in mind, but is a recent development of pure logicians. It is an attempt to devise a symbolism, and rules for the manipulation of the symbols, whereby problems in logic can be set out and solved in a manner akin to that of algebra. (The name "Boolean algebra" is a tribute to the contributions of another English mathematician and logician, George Boole, 1815-1864.) It will be described



Fig. 2. Illustration of De Morgan's theorem

first from the point of view of the logician, and it will then be shown how it can be applied to switching circuits.

First, the symbolism. A suitable way of writing "Both A and B", which is itself an abbreviation of a statement like "It was wet (A) and it was windy (B)", is

AB, (as if multiplying),

and of writing "Either A or B",

A + B, (as if adding).

We can carry this principle further so that, for example. "A and either B or C" is written as

A(B + C)

A stroke through a letter can be used to denote negation. Thus:

A means "not A"

These symbols will suffice for our needs, but now we must have rules for their manipulation. Here are some of them:

(A+C)(B+D)	= AB + AD + CD + CD	Rule 1	
A+AB	- A	Rule 2	
A(A+B)	= A ·	Rule 3	
A+A	= A	Rule 4	
AA	= <b>A</b> .	Rule 5	
A+X .	= 1	Rule 6	
AX	= 0	Rule 7	

Rule 1 tells us that expressions in brackets can be multiplied together as in ordinary algebra. We can demonstrate it if we let A stand for, say Australians, B for Builders, C for Canadians and D for Decorators. Then the equa-tion says (left-hand side) that all the people who are either Australian or Canadian and also either Builders or Decorators comprise the same group as do (right-hand side) all the people who are either Australian Builders or Australian Decorators or Canadian Builders or Canadian Decorators 1

Rule 2 is quite different from normal algebraic practice and is one that often enables one to simplify an expression in algebraic logic. This particular example can

be regarded as an exemplification of the fact that all the people who are either Australians or Australian Builders are no more or less than all the Australians. A similar counterpart can be found for rule 3.

Rules 4 and 5 are readily acceptable. "Either A or A" is the same thing as "A". So is "Both A and A".

Rule 6 expresses the logical principle that "everything must be either A or not A, no matter what A may be". Therefore, a collection of all the things that are either A or not A, i.e., A + A, consists of the entire universe, which we call "1". So A + A = 1.



Fig. 3. Some circuits and their representation in algebraic logic

And lastly, in rule 7,  $A_A = 0$ , we have a statement of the principle that nothing can be both A and not A.

We have already observed that formal logic exhibits a complete generality, so that once we have satisfied ourselves that an argument is valid we can use it to talk about any specific things, be they real or imaginary. This same generality allows us to use the symbols of algebraic logic in describing switching circuits, and to use the theorems or rules in designing them.

An immediate start can be made once we have agreed to represent normally-open and normally-closed relay contacts by A and  $\measuredangle$  (or any other letter as appropriate) respectively, to write parallel branches as if they were added, and series branches as if they were multiplied. Fig. 3 shows some circuits and their representations in algebraic logic. Some authorities prefer to use addition for series branches and multiplication for parallel branches. This makes some of the working more difficult, some easier, but does not affect the general principles. The choice of convention is analogous to the choice between working with resistances or with conductances.

The value of algebraic logic in the design of switching circuits springs from the fact that the rules for manipulation of the symbols enable one in many cases to effect a considerable simplification of an expression in algebraic logic, with an equal simplification of the circuit to which it corresponds.

Suppose that one wishes to simplify the circuit of Fig. 4(a), which embodies contacts of relays, A, B, C and D.

The circuit is expressed in algebraic logic, using our agreed symbolism, as:

AC +  $(\measuredangle + \cancel{p})(B + C) + C(B + \measuredangle D)$ . This equals, on multiplying out the brackets by rule 1,  $(AC + \measuredangle B + \measuredangle C + B\cancel{p} + C\cancel{p} + BC + \measuredangle CD)$ =  $(A + \measuredangle) C + \measuredangle B + B\cancel{p} + C\cancel{p} + BC + \measuredangle CD)$ =  $C + \measuredangle B + B\cancel{p} + C\cancel{p} + BC + \measuredangle CD)$ Since  $A + \measuredangle = 1$ , by rule 6 =  $C + \measuredangle B + B\cancel{p}$ Since  $C + C\cancel{p} + BC + \measuredangle CD = C$ , by rule 2 =  $C + B(\measuredangle + \cancel{p})$ 





No further simplification is possible and so the final expression is converted into circuit form to obtain Fig. 4(b). This uses only four contacts instead of the original ten, but the two circuits are equivalent in the sense that, for any given condition of the relays A, B, C and D, the path from X to Y will be open or closed in Fig. 4(b) according to as it is open or closed in Fig. 4(a).

The circuit designer would have arrived at Fig. 4(b) quite easily without the aid of algebraic logic. This particular example has been given only because it is simple, brings in three of the rules and shows how very effectively this new technique produces the right answer; but if the reader cares to pursue the subject further he will find, in current literature, examples of the use of algebraic logic as an effective method for the development of complex switching circuits.

#### Conclusion

An attempt has been made to show some aspects of that science which stands in the background of algebraic logic, and which is called formal logic. It is hoped that this will have been found interesting, and that the later remarks on the application of algebraic logic to switching circuit design will have given the reader an introduction sufficient to enable him to attack, without hesitation, further treatizes on the subject.

### A Differential Input-Stage for Low-Frequency Amplifiers

By V. H. Attree\*, B.Sc.

An A.C. coupled differential input-stage is described in which the ratio of in-phase to anti-phase gain (i.e., the "transmission-factor") is theoretically infinite. In a practical circuit the measured transmission-factor is 10 000 at an in-phase signal level of 1.0V R.M.S. The frequency response is 0.1-20 000c/s which is sufficient for most applications in engineering and biology.

A LOW-FREQUENCY amplifier is often required to measure a small anti-phase signal in the presence of a large in-phase interfering voltage. The usual type of inputstage for such an amplifier consists of two triodes with a common cathode load<sup>1,2</sup>; one of the requirements for good rejection of in-phase voltages is a high value of cathode resistance which can be obtained by using the A.C. resistance of a pentode valve as the cathode load. An amplifier with this arrangement was described by Johnston<sup>3</sup>. Unfortunately the balance also depends on the difference between the amplification factors ( $\mu$ ) of the two triodes and, as pointed out by Harris and Bishop<sup>4</sup>, this effect may cause the performance to be very much worse than would be expected from an analysis based on the assumption that  $\mu$  was the same for both triodes. The problem was further discussed by Parnum<sup>6</sup> who suggested a new term "transmission-factor" for the ratio between in-phase voltage and the anti-phase signal required to give the same output. Parnum<sup>6</sup> and McFee<sup>6</sup> have both proposed equalizing the u's of the valves by anode to cathode feedback while Davis' employs a potentiometer in the grid of one of the valves. The use of the potentiometer has the objection that it sets a limit on the input impedance and gives an appreciable phase unbalance which will affect the performance. It is important to note that the transmission-factor will in general fall at high signal-levels, due to overloading, and therefore the in-phase voltage at which it is measured should be given.

In the present circuit the signals from the two inputs are fed to the grid and cathode of a triode in such a ratio that, when they are in phase, the triode output is substantially zero. This principle is essentially that of the  $\mu$ bridge stabilizer<sup>8</sup> which is often used for regulating photomultiplier E.H.T. supplies<sup>9</sup>. The application to a differential input-stage appears to have been first described by Conrad<sup>10</sup>.



**Circuit Operation** 

The rejection of in-phase signals is obtained in the manner shown in Fig. 1. One signal is applied to the

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grid of a triode and the other to the negative end of the cathode load. The anode current is:—

 $i_a = (\mu/r_a) [e_2 - e_1 - i_a R_k] + (1/r_a) [e_2 - i_a (R_a + R_k)]$ If  $i_a$  is to be zero then we have the exact relation  $\mu e_1 = (\mu + 1) e_2$  i.e., the balance depends only on  $\mu$  which, in a given valve, is constant over a wide range of anode currents. In the practical circuit the voltages  $e_1$  and  $e_2$  are obtained from cathode-followers: this enables the necessary adjustment of the ratio  $e_1/e_2$  to be made with a low-valued potentiometer.

### Differential Input-Stage with Single-Sided Output

The circuit diagram of the differential input stage is shown in Fig. 2. The two halves,  $V_1$  and  $V_2$  of the 6SN7



Fig. 2. Differential input stage with single-sided output  $R_{15} = O - 2M\Omega$  selected to match components.

provide the cathode and grid signals for the bridge valve  $V_3$ . The required ratio for zero output from  $V_3$  anode, with an in-phase signal to  $V_1$  and  $V_2$ , is set by adjustment of the  $2k\Omega$  potentiometer  $R_{11}$  in  $V_2$  cathode. The anode current of  $V_3$  is 0.3mA while  $V_1$  and  $V_2$  each take 0.7mA. The  $220k\Omega$  resistor  $R_{10}$  provides a current in  $V_3$  cathode resistor  $R_{13}$  equal to the anode current of  $V_3$ ; this equalizes the currents in  $R_2$  and  $R_{13}$  and ensures that the D.C. operating conditions in  $V_1$  and  $V_2$  are similar. The capacitors  $C_1$ ,  $C_2$  and  $C_3$  must have high insulation and low dielectric absorption and should be either 500-1 000V paper-dielectric or, better, of the plastic-film type.

Over the middle part of the frequency range the phasechange in the coupling time-constants, and at  $V_1$  and  $V_2$ cathodes, is negligible and an exact null-balance may be obtained at the output. However, at both high and low frequencies phase-change becomes of importance and it is necessary to provide for phase balancing in order to

maintain a high transmission-factor. One side of the heaters is earthed so that the heater-cathode capacitance of  $V_1$  and  $V_3$  appears across  $R_2$ . The capacitance across  $R_2$  is somewhat smaller than the total across  $R_1$ , which includes the stray capacitance of the metal-cased  $1.0\mu F$ capacitor  $C_s$ . Capacitance balance is obtained by the capacitor  $C_3$ . Capacitance balance is obtained by the 20-40pF trimmer  $C_4$ . When the amplifier is fed with an in-phase signal the voltage across  $R_2$  is  $\mu_3 K$  (where K is a constant) and the voltage at V, grid is  $(\mu_3 + 1)K$ . Thus the current in the 1M $\Omega$  grid leak  $R_9$  is  $K/R_9$ , instead of the much greater current of  $K(\mu_3 + 1)/R_9$  which would be obtained if  $R_9$  were returned to a point of zero A.C. potential. tial. The effective time-constant of the coupling capa-citor  $C_s$  and the leak  $R_s$  is therefore  $(\mu_s + 1)$  times  $C_s R_s$ , or about 20sec. At 10c/s the phase-change in this timeconstant is 1/1 256 radians or 1/20 degree and will be large enough to reduce the transmission factor from better than 10 000 down to about 1 000. Low-frequency phase correction is obtained by the resistor  $R_{15}$  in series with  $V_2$  grid leak  $R_{14}$ ;  $R_{15}$  also corrects for any small difference between the input time-constants. The correction is necessarily approximate, as it assumes that the phase-frequency relation of the two-stage network to V, grid can be made the same as that of the one-stage network to  $V_3$  grid can be inade the same as that of the one-stage network to  $V_3$  cathode. The phase-change in the coupling on  $V_3$  grid is smaller than in the coupling to  $V_1$  and  $V_2$  as the leaks  $R_1$  and  $R_{14}$  are "multiplied" to a lesser extent than  $R_9$ , so that in practice the input stages predominate and a good balance may be obtained over a wide frequency range. It is unnecessary to make  $R_{15}$  a variable control as the required resistance depends mainly on component tolerances and is little affected by valve changes. When setting up . the circuit initially it is convenient to adjust  $R_{11}$  at 500c/s,  $C_4$  at 10kc/s and  $R_{15}$  at 10c/s.

Although the circuit gives almost complete rejection of in-phase signals fed to  $V_1$  and  $V_2$  it provides very little attenuation of disturbances appearing on the 120V H.T. line. This is of no importance when the H.T. is taken from a well-screened dry battery but may be troublesome if it is derived from the supply mains. Where the subsequent amplifier has a balanced input a considerable reduction in supply ripple (about 100 times) may be obtained by connecting one side of the main amplifier to the junction of  $R_4$  and  $R_5$  instead of to earth. The circuit may be balanced for H.T. changes by injecting a small ripple voltage (1-10mV) into the H.T. lead and adjusting  $R_{\rm s}$  for minimum output from the main amplifier.

#### Performace

The gain of the stage for a signal applied to either  $V_1$  or  $V_2$  is about four times. The frequency response is substantially constant from 0.25c/s to 10kc/s; it is 3db down at 0.1c/s and 20kc/s. The transmission-factor is 10000 for an in-phase input of 1.0V R.M.S. so that a signal of 1.0V applied to both grids gives the same output as  $100\,\mu V$  applied to either grid separately. The transmissionfactor falls as the in-phase signal is increased until, at 10V R.M.S. (28V peak-to-peak), it is 1 000. A 10 per cent change in the H.T. voltage (120-108V), or in the heater voltage (6.3-5.7V), have practically no effect on the transmission-factor. Increasing the H.T. impedance to  $100\Omega$  reduces the factor to about 3 000 (at 1  $\cdot$  OV R.M.S.) but the original value of 10 000 or better may be restored by readjustment of  $R_{11}$ . The transmission-factor is uniformly high over the frequency band 5c/s-10kc/s; at very low frequencies it deteriorates by an order of magnitude. This falling off of the transmission-factor is unimportant as the in-phase interfering voltages are nearly always due to the supply mains.

### **Differential Input-Stage with Balanced Output**

The circuit given in Fig. 2 may be modified to the symmetrical arrangement of Fig. 3 which has two bridge



Fig. 3. Differential input stage with balanced output

valves cross-connected between the input cathodefollowers. This gives a balanced output and reduces the effect of ripple from the 120V H.T. line. A similar circuit, but with D.C. connexion, has been described by Van Scoyoc and Warnke<sup>11</sup>: this amplifier did not have adjustment for the ratio of the grid and cathode signals on the bridge values and hence the transmission-factor of the first stage was relatively low. The control on  $V_1$  in Fig. 3 is set to give zero output from  $V_3$  with an in-phase signal to  $V_1$  and  $V_2$  while the similar control on  $V_2$  balances  $V_4$ . The use of two balance controls complicates the setting procedure and for this reason the circuit of Fig. 2 which requires only one control, is usually better in practice, although it gives a single-sided output.

#### Applications

The differential input-stage was originally used in this laboratory for coupling a bimorph crystal microphone to a single-sided A.C. amplifier. The microphone was used in some experiments on the aerodynamic noise associated with a turbulent air-flow. The frequency range of interest was 50-1 000c/s and the input capacitors  $C_1$  and  $C_2$  were  $0.01\mu$ F mica dielectric. For biological work good lowfrequency response is usually required and this is obtained fairly easily in the present circuit. In the particular case of electrocardiograms the desirable low-frequency response has been discussed by Dawson<sup>12</sup>, Katz<sup>13</sup> and by Nickerson<sup>14</sup>. The response should be roughly equivalent to that of a single time-constant of two seconds so that the differential input-stage of Fig. 2 may be used prior to a conventional electrocardiograph in order to improve the discrimination against mains-frequency interference.

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### LETTERS TO THE EDITOR

(We do not hold ourselves responsible for the opinions of our correspondents)

A Direct Reading Thermistor Bridge DEAR SIR,-In Messrs. Pearson and

Benson's article in your February issue, there appear to be several mathematical errors.

(a) assuming  $\delta \ll A$  equation (3) should read :

 $I_{\rm g} =$ 

 $\frac{E\ \delta}{4A(A+R_g)}\left(1+\frac{\delta}{4(A+R_g)},\frac{3A+2R_g}{A}\right)$ Εδ

However, from equation (4) using the authors' figures,

$$\delta = \frac{\partial R}{\partial P} \Delta P = 99.2\Omega$$

where the increase of power, dissipated in

the thermistor,  $\Delta P = 2$  mW. The graph (Fig. 2) shows this to be of the right order. This clearly does not satisfy the condition  $\delta \ll A$  (where A = 300 $\Omega$ ).

(b) The authors' numerical substitu-(b) The authors numerical substruc-tion to obtain the second term of equa-tion (3) gives 0.14, not 0.0014 as stated; moreover, using the correct expression for this term the value would become 0.20. This cannot be neglected compared with 1.

(c) Since  $\delta$  is comparable with A, the change in D.C. power, when R.F. power is applied, must also be taken into consideration. It can be shown from an analysis of the bridge that the D.C. power  $W_{DO}$  dissipated in the thermistor is given by:

$$\frac{A}{4} \left[ \frac{E}{A} + I_{\varepsilon} \left( \frac{3 + 2R}{A} \right) \right] \left[ \frac{E}{A} - I_{\varepsilon} \left( \frac{1 + 2R}{A} \right) \right]$$

where E = voltage across bridge. substituting the authors' values for A and R<sub>g</sub>

$$W_{\rm DC} = 75 \left[ \frac{E}{300} + 6.33 I_{\rm g} \right] \left[ \frac{F}{300} - 4.33 I_{\rm g} \right]$$
.....(A)

For balance 
$$W_{\rm BC} = \frac{1}{(300)^2}$$
.....(B)

From Fig. 2 at 21°C. the power dissi-pated in the thermistor to give a resistance of 300 ohms.

$$W_{DC} = 9.5 \text{mW}.$$
  
hence from equation (B)

E = 3.38 volts. It can be shown that:

$$I_{g} = \frac{E \delta}{4A (A + R_{g}) - (3A + 2R_{g})\delta}$$
$$I_{g} = \frac{E \delta}{960 - 1.9 \delta} \mathbf{m} \mathbf{A}$$

In paragraph (2) it was shown that  $\delta = 99.2\Omega$  for an increase in the total power dissipation in the thermistor of 2mW; therefore assuming E is constant:

$$I_{g} = 434 \mu A$$

Using this value of  $I_g$ , the D.C. power dissipated in the thermistor from equation (A) is 9.88mW, giving an increase in D.C. power of 0.38mW. Hence the R.F. power required to increase the total power by 2mW is 1.62mW, assuming a constant-voltage generator.

$$\frac{I_{g}}{\Delta P} = \frac{E \frac{\partial R}{\partial P}}{4A (A + R_{g})} = 175 \mu \text{A/mW}.$$

where  $\Delta P$  = microwave power.

Putting  $I_{\rm g} = 434\mu A$ , the microwave power would be 2.48mW. The correct value has been shown to be 1.62mW. Thus the total error involved in equation (6) is 53 per cent for the values considered.

> Yours faithfully, W. B. W. ALISON, J. C. E. TAYLOR, Teddington. Middlesex.

### The authors' reply:

DEAR SIR,—We thank Messrs. Alison and Taylor for pointing out that there is an error in our numerical substitution to obtain the second term of equation (3) and also that there is a factor of 2 missing from this term.

The fundamental accuracy of the bridge depends, however, on a practical check and not on equations (5) and (6) which are derived from equation (3).

When the bridge is balanced, with no microwave power in the thermistor, the meter-arm current is zero. Our meter-and-shunt arrangement was calibrated against a sub-standard instrument so that against a sub-standard instrument so that a current of  $300\mu$ A gave full-scale deflexion with  $25\Omega$  of the  $50^\circ$  variable resistor in circuit. When the bridge was built it was checked so that 2mW of microwave power gave a full-scale deflexion on the meter. Thus we have two points on the graph of microwave power against meter deflexion, i.e., the power against meter deflexion, i.e., the origin 0 and point P (Fig. A). We have



Fig. A. Graph of microwave

assumed that between 0 and P the graph is reasonably linear. If Messrs. Alison and Taylor's observations have a great effect then this curve will be markedly non-linear between 0 and P. Also, if it is, the bridge must be calibrated for the full range of the meter scale before use.

We have, therefore, checked our bridge theoretically for non-linearity by using equations (12) and (13) to find  $e_t$  and  $i_t$ for various values of  $I_s$  (assuming a con-stant temperature) and then putting  $e_t$ 

and  $i_t$  into equation (11) to find the resulting microwave power  $P_t$ . In this way we can determine whether  $P_t$  is proportional to  $I_g$  and if not then the extent of the non-linearity. Fig. B shows the result of these calculations assuming the temperature to be 21°C. It will be seen that the graph is not absolutely linear but the largest deviation, assum-ing points 0 and P are correct is around  $I_{\mu} = 150\mu A$  and amounts to about 4 per cent.



Fig. B. Result of calculations

It should be remembered that the bridge is not intended to be an accurate laboratory instrument but one for use as a portable power-checking device where absolute accuracy, although desirable, is not essential. It could be used as an accurate instrument, however, if it was calibrated all the way up the scale instead

of just at one point. In Messrs. Alison and Taylor's calculations given under section (c) it is assumed that there is a constant-voltage generator across the bridge and they ignore the fact that the bridge calibration was checked practically when a micro-wave power of 2mW was being dissi-pated in the thermistor. In our final design the voltage actually applied to the bridge is far from constant because of rand D.

We should like to take this opportunity of mentioning that there are two misprints in the article :-

(1) the voltage across the bridge should be 105V as stated on page 56 and not 150V as shown on Fig. 11.

(2) E' in equations (12) and (13) should be defined as E/(1+r/D) and not E/(1+r/R) as given on page 54.

Yours faithfully,

R. M. PEARSON, F. A. BENSON,

Department of Electrical Engineering, University of Sheffield.

### **Corona Discharge Tubes**

DEAR SIR,—The article in your January issue by Mr. E. E. Shelton and Mr. F. Wade is interesting in connexion with the use of discharge tubes for interstage coupling in D.C. amplifiers. It has been pointed out by Miller<sup>1</sup> that the ordinary voltage stabilizer tube will function as a constant-voltage device for currents as small as 10 or 20µA, although the manufacturers' data on the tubes he used states that a current of 5mA is necessary to ensure stable operation.

The description in your January issue of the different types of discharge appears to explain the discrepancy. When the tubes are operated at 10 or  $20\mu$ A a corona discharge presumably occurs, and larger currents produce the cathode-glow discharge. A substantially constant voltage over a range of current values results from either type of discharge, but we should expect to find a range of current somewhere between  $20\mu$ A and 5mA over which the voltage falls sharply with increasing current.

I have made a few tests on stabilizer tubes which tend to support these conclusions. Some tests were made with VR105 tubes since these have been used in D.C. amplifiers by Harris<sup>2</sup> and others. A used tube of this type, of American manufacture, had the expected characteristics. The voltage across it was approximately the ignition voltage from zero current to about 0.3mA. At this current a bright spot appeared on the cathode (at lower currents a faint glow was seen near the anode) and the voltage fell to a value in the region of the normal running voltage.

On the other hand, two new tubes type VR105, of British manufacture, were found to have an ignition voltage very close to the running voltage, and no abrupt voltage change was observed between zero current and the normal operating current. In these tubes it was not possible to observe the discharge as the mica end plates were not sufficiently transparent.

The expected behaviour was shown very clearly by tubes, type 7475, in which the critical current for the change from corona to cathode-glow discharge was about  $10\mu$ A. It is clear that when discharge tubes are used either as stabilizers or in D.C. amplifiers, the operating current should not be allowed to approach the critical value, though it may be either higher or lower.

In my letter which you published in your issue for August, 1952. I referred to the lack of information about the A.C. impedance of stabilizer tubes at high frequencies. I have since discovered that Iannone and Baller<sup>s</sup> give information about the impedance of the VR75, VR90, VR105 and VR150 at frequencies up to 20kc/s.

> Yours faithfully, A. M. ANDREW, The University. Glasgow.

### REFERENCES

- MILLER, S. E., Sensitive D.C. Amplifier with A.C. Operation, *Electronics*, p. 27 (Nov., 1941).
   HARRIS, E. J., An Electrometer Impedance Convertor, *Electronic Engng.* 23, 109 (1951).
   IANNONE, F., BALLER, H., Gas-tube Coupling for D.C. Amplifiers, *Electronics* p. 106 (Oct., 1946).

### The authors' reply :

DEAR SIR,-Dr. Andrew's letter suggesting the application of corona discharge tubes as interstage couplings in D.C. amplifiers is of considerable interest. We would, however, like to point out that some difficulty may be encountered due to the generation of noise in the tube.

The bright spot to which Dr. Andrew refers in a VR105 operating at a current of 0.3mA, would appear to be caused by a sharp point giving rise to a local glow discharge prior to the spread of the glow discharge over the whole cathode surface.

### Yours faithfully, E. E. SHELTON.

F. WADE,

Hounslow, Middlesex.



Source and impedance  $300\Omega$ 

The curves of Figs. 2 and 3 show the results obtained for both paths. In practice, the coupling capacitor between the first valve and the filter determines the frequency at which maximum cancel-



Fig. 1. The complete circuit

### A Selector Rejector Circuit for a 3.2kc/s Pilot Signal

DEAR SIR,—Where a fixed frequency pilot signal is used in a transmission scheme for assessing the correct operation of apparatus, the need often arises to measure the level of the pilot signal and also to remove it from the transmitted frequency spectrum. This can be accomplished with a band-stop and a band-pass filter, however, in the circuit about to be described, a single frequency rejection filter is used for both operations.

Fig. 1 shows the complete circuit. The Fig. 1 shows the complete circuit. The incoming signal, that is, speech plus a 3·2kc/s tone is applied to the control grid of an amplifying valve and the primary winding of a transformer. A bridged-T rejection filter in the anode circuit of this valve removes the pilot tone from the speech signal and this is tone from the speech signal and this is available, through a resistance pad, for

audible monitoring or re-transmission. The secondary winding of the input transformer supplies a signal out of phase and equal in amplitude to the output of the rejection filter. These signals are combined in another resistance pad and applied to the control grid of a triode connected stage for measurement by a rectifier meter. Thus, at frequencies other than the rejection frequency of the filter, cancellation occurs, and no signal is presented to the control grid of the triode. At the rejection frequency, triode. At the rejection however, only the signal from the input transformer is available and this measured by the meter. is' lation takes place and hence, the rapid-ity of cut-off from the frequency of peak meter reading. The capacitor in this case provides maximum cancellation at about 400c/s and the potentiometer across the secondary of the transformer assures equality of amplitude.



#### Fig. 3. Frequency characteristic of meter path Input level 16.4 db REL IV. Source impedance 300 Q

A centre tapped inductance of 1.8H was used in the filter, the rejection fre-quency being adjusted by the trimmer capacitor and the balance by the parallel arm variable resistor. Initial calibration of the meter could be effected by a variable bias resistor in the triode stage. The transformer used had a ratio of 1:16 and a primary inductance of 2H at 50c/s.

### Yours faithfully,

#### E. J. MILLER,

P.O. Research Station.

### Strain Gauges Theory and Application

By J. J. Koch, R. G. Boiten, A. L. Biermazz, G. P. Roszbach and G. W. Van Santen. 101 pp., 26 figs. Demy 8vo. Philips Technical Library. Elsevier Press Inc., New York, Cleaver Hume Press Ltd., London. 1952. Price 16s.

THIS little book has been produced "with the intention of providing . a concise but comprehensive description of the technique of measuring with tion of the technique of measuring with strain gauges and everything appertain-ing thereto". The authors jointly have clearly had a great deal of practical experience in this field, and what they have to say is mostly accurate and to the point. It is a pity that not all the techniques are described in sufficient detail, and that the accuracy of the information is not maintained through-out. For example, on page 80 a method out. For example, on page 80 a method is given of constructing Mohr's strain circle from rosette data, and this is quite erroneous. The careful reader will be erroneous. The careful reader will be able to make the necessary change in the construction, but the need for this is unfortunate.

A general criticism of this book is that it refers with very few exceptions to gauges and apparatus manufactured by Philips Industries, and in places reads like a sales leaflet rather than a book of technical interest. The illustrations frequently have captions which refer to a Philips catalogue number, and a search in the text is required to amplify the caption. There is no bibliography or detailed index.

The chapter on the technique of cementing and connecting strain gauges is perhaps the best, the author having drawn on an intimate knowledge of the necessity for careful preparation. That dealing with the theory of elastic deformation is unduly elaborate, as it begins with the 3-dimensional case and later simplifies the theory to describe the 2-dimensional stress field. Since strain gauges can act in two dimensions only, this might have been done at the outset. A chapter on measuring instruments employing strain gauges is interesting. No effective discussion of slip ring problems appears.

### J. YARNELL

### **Electronic Analogue Computors**

By Granino A. Korn and Theresa M. Korn. 378 pp., 181 figs. Royal 8vo. McGraw-Hill Book Co. Inc. 1952. Price 59s. 6d.

THIS book deals with D.C. electronic analogue computors suitable for use as differential analysers and simultaneous equation solvers. The basic computing equation solvers. The basic computing elements are adding devices (summing amplifiers), multipliers, integrators, and special function generators such as sine-law potentiometers. (Differentiators present certain practical snags, and fortu-nately it is usually possible to arrange computations so that their use is not necessary). The first chapter is a sum-mary of the whole book, is excellently written, and can be thoroughly recom-mended to anyone wishing to obtain an overall picture of the principles, prac-tices, and powers of this type of computor without going into much detail. detail.

The dual authorship is clear from the text. Those chapters which deal with general principles and with the use of

the machines are written clearly and go into considerable detail. They will be of great value to those who use or hope to use such computors. However, many readers might be rather dismayed by the freedom with which the operator P, defined as d/dr (r being time), is used, particularly in Chapters 4 and 5, and further explanation of this operator is

BOOK REVIEWS

desirable. The chapters dealing with the design of the computing elements, which are of more interest to electronic engineers, are less satisfactory. The standard of the electronics presented is curiously uneven. For example, in Chapter 5 the elementary principles of the design of an amplifying stage are given, whereas in the same chapter there is a discussion of the stabilization of feedback amplifiers which will be beyond the mathematical ability of many competent practical engineers.

There is also some lack of performance figures. For example, there is nowhere any indication of the order of time-constants obtainable with feedback integrators. The accuracies obtainable with some of the other types of computing elements are, however, given.

One or two other criticisms may be briefly mentioned. The circuit diagrams are not well drawn on the whole; but this tends to be a common failing with American publications. More use might have been made of equivalent circuits, and more mention should have been made of photo-electric curve-followers, which are in fairly common use as computor input devices. In chapter 4 the authors imply that the parallel feedback integrator operates by increasing the effective value of the charging resistor, whereas a sounder physical resistor, whereas a sounder physical explanation is that it increases the effective value of the capacitor.

There is a fairly extensive bibliography,

the far greater part of which refers to works published in the U.S.A. To sum up, the book is well produced, will be useful to users of electronic analogue computors, and will possibly be of interest to electronic engineers.

> M. J. TUCKER .

### Storage Tubes and their Basic **Principles**

By M. Knoll and B. Kazan. 143 pp., 34 figs. Demy 8vo. J. Wiley & Sons Inc., New York. Chapman & Hall Ltd., London. 1952. Price 24s. THE phenomenon of electronic storage "THE phenomenon of electronic storage in television, computor and signal-convertor applications is nowadays a well recognized principle, but storage tubes themselves have not been described in the literature in any great detail. This book consists, to a large extent, of information which was originally pre-pared as a report for the technical branch of the US Army Signal Corps. The of the U.S. Army Signal Corps. The authors have not described in detail the construction of storage tubes, but have concentrated on the fundamental storage principles.

While the subject matter of the book is dealt with in what is obviously an authoritative manner the layout and type setting are unorthodox and curious to a degree. The use of a mixture of Roman and Arabic numerals, capital and a degree. small letters (carried in an extreme case to such limits as-Part III.A.5.a) can no doubt be attributed to the parentage of the book mentioned above, but it appears to the reviewer to be excessive filial devotion to deny the reader at least the good looks of justification of the printed lines. There is also an odd economy of type face, which seems to add unnecessarily to the severity of the book.

Part I deals with the potentials acquired by an insulated target element under a steady electron bombardment, an understanding of which is essential to any analysis of a charge pattern in a storage tube. Curves of equilibrium voltages and potential shifts are given and the significance of their precise shapes is discussed in detail.

Part II comprises a short chapter setting out the definitions of the terms to be used in the remainder of the volume. (The convention of alphabetical listing is ignored !)

betical listing is ignored !) Part III is of particular importance to the newcomer to the subject, as it discusses very thoroughly the various methods of writing, i.e., forming the charge pattern on the target, and read-ing, i.e., generating the output signal from the stored information, by control-ling secondary emission and bombard-ment conductivity. The processes of ment conductivity. The processes of photoemission and photoconductivity which are of importance in television camera tubes are also discussed.

Parts IV to VII are devoted to the discussion of various types of storage tubes. Signal-convertor tubes such as the Pensak Graphechon and the Haeff Memory tube are dealt with in Part IV. Viewing storage tubes in which the output signal is obtained as a visual output signal is obtained as a visual signal on a phosphor form the subject of Part V. The cathode-modulation tubes of this general type may have applications in future developments of television and in radar operation. Part VI deals with computor storage

Part VI deals with computer storage-tubes which are designed to store Yes-No information for very long periods. Part VII treats with storage tubes of the television camera type. The well

the television camera type. The well known Iconoscope, Orthicon, Image Iconoscope and Image Orthicon tubes are described as well as the newer photoconductive Vidicon and the scattered electron Image Isocon.

The book is completed by a useful and comprehensive bibliography which forms a most valuable and important section of the book. References are by no means limited to work carried out in the U.S.A. In particular the references to the various: tubes discussed in Parts IV to VII will put the reader on the track of the many design features which are (intentionally) omitted from the present volume. Many of the tubes described in the

book are still of a developmental nature and the authors have made no attempt to deal with their practical applications. The subject matter of the book is so specialized that it is not to be expected that it will be of wide appeal, but it is sure to be welcomed by all tube designers and those workers in allied fields, whose need for a suitable storage tube often exceeds their knowledge of the complex theories involved. It is in essence a concise collection of theoretical data which has not before been collected together and within the limits explained in this review, it can be said to achieve its authors' aims.

R. E. B. HICKMAN.

### **Electron Tubes in Industry**

By Keith Henney and James D. Fahnestock. 3rd Edition. 345 pp., 196 figs. Royal Svo. McGraw-Hill Publishing Co., Ltd. 1952. Price 51s.

**Disasting Co.** Let 1322. The same **Disastration** of theoretical dis-cussions in this book, which is intended to satisfy the needs of non-technical industrial personnel who come into contact with electronic aids to industry. contact with electronic aids to industry. The allegations made by the authors concerning the Foster Seely fequency discriminator form a masterpiece of muddled misconception: "Each diode gets equal voltages so long as the incoming frequency is equal to the resonant frequency . . . At other fre-quencies the diode input voltages are equal but out of phase with the input to the transformer. The diode output to the transformer. The diode output voltages are then unequal"! The voltages are then unequal if the essential vector diagrams which would have provided a true picture are omitted. On page 13, the sketch of a parallel resonant circuit shows the resistance in the supply lead to the tuned circuit instead of in the inductive branch. Only forth particular the particular a simple a fleeting reference, with neither a simple theory nor a circuit diagram, is made to the blocking oscillator. Phase shift RC oscillators, crystal controlled oscillators, oscillators, crystal controlled oscillators, the transitron, Hartley and Colpitts oscillators are conspicuous by their absence. The sketches on pages 3, 6, and 10, which are poor apologies, differing among themselves, for sinusoidal waveforms, are quite inexcusable in a 3rd edition. The cathode-follower is mentioned, but no reasons are given for its basic properties, and its limitations are its Dasic properties, and its initiations are ignored completely. The simple copper "slugged" relay is not included in the methods of delaying the operation of electromagnetically operated relays. In the otherwise detailed description of the telephone relay, no mention is made of the essential stud residual or of the reason for its existence. Typical of the looseness of expression and lack of looseness of expression and lack of precision which is rampant is the state-ment on page 126 that the efficiency of a typical R.F. E.H.T. unit is "fairly good". The authors have succeeded in produc-ing a hybrid of "electronics in three easy lessons" and a review of current

The authors have succeeded in producing a hybrid of "electronics in three easy lessons" and a review of current practice, a hybrid that is neither flesh, fowl, nor good red herring. Although the seasoned engineer may, with due caution, be able to use parts of the book for reference, the newcomer to the art of electronics is advised not to delve into it. Most certainly, it falls short of the publishers usual high standard.

H. STIBBE.

### By A. G. Lyle. 355 pp., 111 figs. Demy 8vo. Chapman & Hall Ltd. 1952. Price 45s. A S the author states in his preface, this

As the author states in his particular book is written primarily for the practising engineer, and as such would seem to succeed admirably, even if only for the mass of valuable data and information compressed between its covers. Judged upon the quality of the said data, first impressions are fully confirmed. The emphasis is throughout on the practical rather than the theoretical aspect of the subject, as is of course only right in a book of this nature.

Major Faults on Power Systems

The first few chapters are devoted to the general subject of faults and their calculation. The book then goes on to deal in detail with protective systems (including proprietary ones), oil and air circuit breakers, concluding with a section on the design of switchgear, and neutral inversion. The final chapter is devoted to conclusions based on a multitude of test results. Several appendices of calculations and an analysis and discussion of circuit breaker test results are included, illustrated with oscillograms and photographs. A short bibliography is included at the end of each chapter.

The whole book is very adequately illustrated with diagrams, photographs and oscillographic records. In addition to the switch and protective gear specialist, the advanced student will find parts of the book well worthy of his attention. G. H. PLATT.

### Higher Industrial Production with Electricity

145 pp., 104 figs. Medium 8vo. British Electrical Development Association. 1953. Price 9s.

THE whole subject of higher industrial productivity is an extremely wide one, but this book concentrates on the more striking examples of electrical methods, and should be of particular interest to industrialists as providing a review of the many ways in which modern applications of electricity have increased productivity. This volume is introductory to the Electrial Development Association's "Electricity and Productivity" series of eight books.

### Electronics for Communication Engineers

By John Markus and Vin Zeluff. 610 pp., 758 figs. Crown 8vo. McGraw-Hill Publishing Co., Ltd. 1952. Price 85s.

THIS book is composed of 252 articles which have appeared in *Electronics*. Topics are arranged by subject in sixteen chapters, including such subjects as amplifiers, components, filters, oscillators, power supplies, and transmission lines. It gives selected design articles with many equations, charts, and nomographs. There are ten articles on electronic music, giving design data and circuit details of both commercial and custom-built electronic organs of various types. The chapter arrangement is alphabetically by subject, and indexing by title, author, and cross index of subjects all facilitate the reading of this book.



### ELECTROPHYSIOLOGICAL TECHNIQUE

By C. J. Dickinson, B.A., B.Sc. (Magdalen College, Oxford)

Price 12/6

The author describes the use of electronic methods as applied to research in Neurophysiology. Chapters are devoted to amplifying, recording and stimulating techniques used in physiology and medicine (e.g. electrocardiography, electroencephalography, etc.)

Order your copy through your bookseller or direct from



### ELECTRONIC EQUIPMENT

A description, compiled from information supplied by the manufacturers, of new components, accessories and test instruments.

### A V.L.F. Amplifier

THIS amplifier has been developed to provide a power source for transducers being used for vibration and fatigue tests at frequencies below the range of a standard audio amplifier.

All components, with the exception of the output transformer, are mounted on vertical panels and contained on a 78in. rack frame. The side and back covers are removable for inspection and maintenance of wiring and components, while the valves are accessible from the front of the rack.

Flexible cables connect the rack to the output transformer which is mounted in a separate cabinet. Wheels are fitted to make this heavy component more easily moveable.

The circuit arrangement consists of a paraphase 3-stage voltage amplifier feeding a grid driven, push-pull, triode output



stage. 15db of negative feedback is applied over 3 stages from the output to the cathode of the second stage. These feedback arrangements are such as to permit the connexion of either highly inductive or capacitive loads to the output transformer without upsetting the stability of the amplifier. However, with a purely reactive load, the maximum permissible output is 350VA. If this is insufficient to drive the transflucer, the power factor of the load can be corrected to unity by the insertion of inductance or capacitance and the full 1 000 watts utilized.

The diagram above shows the maximum permissible output of the amplifier into a "matched" resistive load over the frequency range 2c/s-4000c/s. The lower curve gives the inherent frequency response of the amplifier. The distortion at 1kW output is less than 5 per cent between 10c/s and 1000c/s. The output transformer secondary is

The output transformer secondary is wound in four sections of 50 volts each. These windings may be connected in series, parallel or series-parallel to match to load impedances of 2.5. 10 or 40 $\Omega$ . The input impedance of the amplifier is 600 $\Omega$  free of "earth." A centre-tap

The input impedance of the amplifier is  $600\Omega$  free of "earth." A centre-tap on the input transformer primary permits "balanced to earth" working or, alternatively, one side can be "earthed" for connexion to an unbalanced input source. An input of 0.05V is required for full output. A gain control is included ahead of the 1<sup>st</sup> stage and a selector switch and meter provide means for checking the anode current in all valves in the voltage amplifier. The output valves have their own individual anode current meters.

At the side of the main On/Off switch is a 3-position rotary switch which either operates H.T. locally or allows remote operation.

W. Bryan Savage Ltd., Westmoreland Road, London, N.W.9.

### Industrial Counting Equipment (*Illustrated below*)

THE electronic counter, Type D.2, is a general-purpose unit for application to presses, automatic machinery and conveyors when the products to be counted are partially or entire'y metallic and either ferrous or non-ferrous.



An installation comprises the detector unit, complete with 6-figure electromagnetic counter and a pick-up coil which is located at the point of counting. Components, such as turned parts and small pressings have merely to be dropped through the coil unit aperture, a count being registered as each component passes. Standard interchangeable coil units are available for components varying in size from, say, a 2B.A. screw up to 2in. diameter. The unit illustrated has a 2in. aperture.

For larger objects and for counting on conveyors, the coil unit is so positioned that the articles fall or move past the aperture within approximately 1in. This distance, which varies with the size of the article, is independent of the thickness of the metal and objects wrapped in tinfoil, for instance, are readily counted. Cardboard and other non-metallic containers may also be counted when they have either a tinfoil label or are metalbanded.

The equipment is suitable for 200-250 volts A.C. and is independent of supply variations.

#### Teledictor Ltd., 214 Birmingham Road, Dudley.

An Electronic Temperature Controller IN a number of physical, chemical and biological laboratories, means are required for the precise measurement and control of temperatures, especially of water and oil baths. To meet such needs the Equipment Division of Mullard, Ltd., have recently made available a temperature controller, type E.7594. This instrument enables the temperature of water baths and similar apparatus to be controlled to within fine limits, the response time being very rapid. The instrument has the additional advantages of being simple in operation, and compact and robust in construction.

The sensitive element in the new temperature controller consists of a temperature-sensitive resistor, designed for direct immersion into the fluid to be controlled. This resistor forms one arm of an A.C. bridge network. The output from this network is amplified and is used to control a phase-sensitive circuit operating a relay. Since the phase of the bridge out-



put changes very rapidly at the point of balance, quick action of the relay is ensured.

The relay in the instrument is of the single-pole, double-throw type. This can switch up to 1A at 230V, so that most heating elements used in water baths can be effectively operated.

The instrument has a temperature range of 15-75°C. covered in three steps. A large dial allows the setting point to be precisely located and locked in position. Although the calibrated accuracy of the scale is  $\pm 1$ °C, it is possible to control the temperature to within  $\pm 0.2$ °C about the working point.

A four-position range switch is provided which alters the value of the resistor in the fixed or standard arm of the bridge circuit. The adjacent arm of the bridge contains the temperature-sensitive element, while the remaining two arms are provided by the potentiometer balance control. The fourth position of the range switch connects the bridge to a socket on the front panel of the instrument. This enables a precision external resistor to be connected into the bridge in place of the internally-connected standard resistor. By keeping the external resistor at a constant temperature, a greater accuracy of control can be effected. Moreover, through the choice of suitable resistance values, the range of the instrument can be extended to cover any temperature between minus 50°C and plus 80°C. At the lower position, however, the range covered by the fine control will be only 10°C.

Mullard, Ltd., Century House, Shaftesbury Avenue, London, W.C.2.

### Humidity Measurement and Control (Illustrated below)

THE Humicon was designed to meet an Inter-Service Specification and W. H. Sanders (Electronics), Ltd., have now been granted a licence to manufacture and supply this instrument to other users.

The detector head consists of glass fibre placed between two rhodium plated electrodes. The electrodes are



perforated to allow a free circulation of air through the glass fibre and are spaced apart by a ring shaped glazed porcelain insulator at the edges. Due to the affinity of very fine fibre glass with airborne moisture and the fact that a number of fibres are placed in close proximity, the moisture will flow between the fibres due to capillary action. Over the range 42 per cent to 100 per cent relative humidity the resistance of the detector varies from  $100M\Omega$  to  $4000\Omega\Omega$ . The detector head contains no perishable material and is capable of operating throughout the temperature range of  $-40^{\circ}C$  to  $+80^{\circ}C$  without serious change in the response curve form or position.

The control unit allows an alternating current to flow across the detector head. A.C. is used to eliminate any polarizing effects which might occur with D.C. The circuit itself consists of an A.C.

The circuit itself consists of an A.C. bridge measuring arrangement, which incorporates three grid controlled hot cathode gas discharge tubes. Also incorporated are three relays from which humidity control apparatus can be operated.

W. H. Sanders (Electronics), Ltd., 48 Dover Street, London, W.1.

### A High-Speed Counter

(Illustrated below) ·

THE Boulton Paul P181 high-speed counter serves two primary functions: (a) The instrument includes a 1000c/s frequency source which, though simple, is nevertheless highly stable for long periods and can readily and rapidly be corrected for drift, caused by ageing of valves and components, by the use of a pre-set trimmer control, checked by measuring the count over a known time interval.

When the counter is connected to this oscillator the instrument may be used for measuring time intervals between two events to an accuracy of  $\pm 5$  milliseconds. The maximum interval counted directly is 10 seconds, but greater intervals can be measured by noting the number of times the one-second counter completes ten steps.

The commencement and completion of



a counting interval are controlled by high-speed relays, which are actuated either by means of push-button switches on the unit or, alternatively, by means of remote switches connected to relevant points on the system under test.

of remote switches connected to relevant points on the system under test. (b) For the purpose of counting pulses of either regular or irregular periodicity over a known interval of time, the internal frequency source is disconnected. Pulses derived from the system under test are fed into the unit, the counting interval being controlled, as before, by means of push-button switches or by remote auxiliary timing switches.

The maximum rate of counting is 1 000 pulses per second and the maximum direct count is 10 000 digits, although, as previously explained, a higher count can be obtained by observing the revolutions of the final indicator.

The instrument may be used as an R.P.M. indicator by coupling a detector element to the rotating shaft and noting the count over a fixed time interval. At 6 000 R.P.M. the speed would normally be checked over an interval of one or two minutes, the only error involved being the time error in synchronizing a stop-watch to the start and stop switches. If reduced accuracy is acceptable the speed can be measured over a proportionately shorter time interval.

Boulton Paul Aircraft Ltd, Wolverhampton.

#### A Conductive Ceramic for Microwave Applications

#### (Illustrated below)

In waveguides, dummy loads are required to absorb the radiation without appreciable reflexion.

Previously, such loads have consisted of conductive particles embedded in an insulating matrix, for example carbon in asbestos, or such materials as wood which are poor insulators. These materials, however, suffer from the disadvantage that, when high local temperatures occur, due to the absorption of high powers, their electrical properties alter, and, in some cases. they become distorted. To overcome this the conductive ceramic "Caslode," has recently been developed by the Plessey Company, Ltd. A white, homogenous material, it is capable of absorbing power at centimetre and millimetre wavelengths, and can thus be used effectively for high power dummy load and attenuator pads in waveguide applications. Caslode is a true semi-conductor, and does not contain dispersed conductive particles. At 3-2cm wavelength its dielectric constant is about 20, and the loss angle 10° to 20°.

During manufacture the material is fired at a high temperature to form a homogenous structure which is stable under extreme conditions of temperature and humidity. Its thermal shock pro-



perties are excellent, and its coefficient of linear expansion is  $6 \times 10^{-6}$  °C. Mechanically, Caslode is similar to

Mechanically, Caslode is similar to porcelain, but can be ground and shaped rather more readily. The raw materials are formed into the required shape before firing and precise dimensions and a good surface finish are subsequently obtained by wet grinding.

by wet grinding. High power dummy loads of this material have been developed to meet the demand for a matched waveguide termination capable of absorbing high powers without the necessity for water cooling. Wedge shaped components have been designed to operate in the common sizes of waveguides. Recent measurements have indicated that units originally developed to give a match of about 0.95 and to dissipate about 100 watts mean power at 3cm wavelength, can absorb 250 watts mean power without any appreciable change in matching properties.

Where a load must occupy the minimum possible space Caslode can be formed into short wedges (e.g. 5cm long for X-band waveguides) which give a match better than 0.85 and an attenuation of the order of 30db.

match better than 0.55 and an anchoa tion of the order of 30db. The Caslode attenuator pad is still in the experimental stage, but units of the slot insertion type have been produced, having an attenuation figure of 28db with a match of 0.95. These units have been tested up to a mean level of 15 watts dissipation.

The Plessey Co., Ltd., Ilford, Essex.

### Notes from the Industry

The President of the Institution of Electrical Engineers, Colonel B. H. Leeson, O.B.E., T.D., at a recent meeting presented the Certificate of Honorary Membership to Sir Harry Railing for his services to the electrical engineering profession and to science. The Faraday Medal was presented to Colonel Sir A. Stanley Angwin for his contributions to the development of telecommunications in Great Britain and in the international and intercontinental fields.

The Radio Corporation of America announces that a talk on the subject of transistor circuits and applications will be given by Mr. G. C. Sziklai of the RCA Research Laboratories, Princeton. Dr. R. L. Smith-Rose has consented to act as chairman of the meeting, which will be held in the Royal Society of Arts lecture theatre at John Adam Street, London, W.C.2, on Wednesday, 1 July, 1953, at 5.30 p.m. Those interested in attending are invited to write to RCA in advance at The Tower, Brook Green Road, London, W.6.

The Ampere Medal, presented by the Societé Française des Electriciens of France has been awarded this year to Monsieur H. Andre. This medal is awarded in consideration of a life's work in some branch of the electrical field. Monsieur H. Andre is the inventor of the modern form of silver-zinc accumulator, the Commonwealth rights of which are held by Venner Ltd.

British Insulated Callender's Cables, Ltd., announce that their Overseas Reception Office has been transferred to 11 Bedford Square, London, W.C.1, telephone Museum 1600.

Rees Mace Marine, Ltd., marine subsidiary of Pye, Ltd., Cambridge, have received an order from the Dover Harbour Board to equip the Port of Dover with V.H.F. radio-telephones for port control. Engineers from Business Radio (Southampton), Ltd., the Pye associated company, are fitting several new companies in the Southampton area with V.H.F.

Mr. C. L. G. Fairfield, Barrister at Law, has joined the Telegraph Construction and Maintenance Co., Ltd., as manager of the Overseas Division.

The BBC announce that mediumpower standby transmitters having an output of 5kW vision and 2kW sound, have now been installed at the Sutton Coldfield television transmitting station. Switching is provided so that they can be used with either the main or reserve aerial systems.

20th Century Electronics, Ltd., have opened a new factory at Centronics Works, King. Henry's Drive, New Addington, Surrey, for the manufacture of Geiger counter tubes. Mr. N. B. Balaam has been appointed general manager, and Mr. H. W. King, production manager.

The Minister of Supply (Mr. Duncan

Sandys) has appointed Mr. S. Scott Hall, C.B., to be Head of Technical Services, British Joint Services Mission, Washington, in succession to Sir Alwyn Crow, C.B.E.

Pye, Ltd., in competition with several other manufacturers, including a number of American firms, have secured a substantial television contract for Radio Tokyo. During the summer three camera chains and the associated control equipment, all of which will operate on the American system, will be shipped to Japan.

The Directors of the Engineering Centre, Ltd., Glasgow, have resolved to close the Centre at the end of September. Certain industries, trade associations and manufacturers have shown preference for arranging their own exhibitions of their products in areas of their own choice.

The BBC have announced with regret the death of Mr, Simon Orde, engineering information department manager. Mr. Orde's work made him widely known outside the BBC, both in this country and overseas.

Standard Telephones and Cables Ltd. have announced with regret the death of Mr. Charles Eve, a former director of the company and a director of Kolster-Brandes, Ltd.

Dr. H. R. L. Lamont has recently joined the Radio Corporation of America as a member of the scientific staff of the European Technical Representative with headquarters in London.

The International Congress of Mathematicians will be held in Amsterdam from 2-9 September next year under the auspices of the Mathematical Society of the Netherlands. The Congress will be open to mathematicians from all parts of the world. Particulars may be obtained from the Secretariat, 2d Boerhaavestraat 49, Amsterdam.

Emitron Television, Ltd., announce that Radio Italiana of Turin have ordered two Flying Spot Film Channels for their new television station.

**Exports of British radio equipment** of all kinds in March were valued at £2 221 395, which was approximately £180,000 in excess of the average monthly value in 1952.

Vidor-Burndept's recently established Press Office is situated at 18 Abbey House, Victoria Street, London, S.W.1, telephone Abbey 1328.

Mr. L. Bainbridge-Bell, a contributor to ELECTRONIC ENGINEERING, will be retiring from Admiralty service this summer. Readers will be aware of his work to increase the clarity of electronic circuit diagrams.

Errata. In Letters to the Editor in the May issue, correspondence between Messrs. Simmons and Vale and Mr. D. T. R. Dighton, p.218, "Strains of 10-°in." should read "strains of 10-°in."

### PUBLICATIONS RECEIVED

EDISWAN MAZDA RADIO AND TELEVISION VALVES AND CATHODE-RAY TUBES is a folder produced by the Edison Swan Electric Co., Ltd., 155 Charing Cross Road, London, W.C.2. and contains information on valves, cathode-ray tubes, E.H.T. rectifiers, etc. As a separate publication there is an index and price list of the various valves and tubes mentioned.

vanous valves and tubes mentioned. EDISWAN GERMANIUM CRYSTAL RECTI-FIERS is a leaflet on germanium crystal diodes which are useful to radio and television set designers as they occupy a relatively small space, are self-supporting in wiring, have negligible capacitance, and require no heater power or socket. Two types suitable for television receiver applications are now available. Edison Swan Electric Co., Ltd., 155 Charing Cross Road, London, W.C.2.

London, W.C.2. PHILIPS ELECTRONIC MEASURING INSTRU-MENTS is a leaflet displaying the wide range of electronic apparatus which this firm offer in the way of test gear required by engineers to meet increasingly stringent requirements. These instruments cover both the radio and the more recent television fields where pattern generators, valve voltmeters and oscilloscopes are used. Philips Electrical Ltd., Century House, Shaftesbury Avenue, London, W.C.2. STANDARD TELEPHONES AND CABLES

Avenue, London, W.C.2. STANDARD TELEPHONES AND CABLES have recently produced a general guide to capacitors. This guide features the standard forms and cases available, and the capacitances and voltages which can be accommodated for various ranges of working temperature. Important factors to the equipment designer are the capacitances and voltages that can be accommodated in a given container, and this guide is presented to assist such engineers. A brochure on suppressors produced by this company is also available. Standard Telephones and Cables Limited, Connaught House, Aldwych, London, W.C.2. THE BRUSH FLECTRICAL ENGINEERING

THE BRUSH ELECTRICAL ENGINEERING CO, LTD., present a brochure on a range of high rupturing capacity fuse links on which they have recently completed development. Details of construction, short circuit performance, protection of motor circuits, dimensions and prices are given. The Brush Electrical Engineering Co., Ltd... Hopkinson Works, Cardiff.

Physical properties of Phoenix INDUSTRIAL GLASS is a booklet describing the physical data on the many useful properties of Phoenix glass. Its thermal properties, mechanical strength, electrical and optical properties, chemical stability, etc., are mentioned. The British Heat Resisting Glass Co., Limited. Phoenix Works, Loxdale Street. Bilston, Staffs. THE SCIENTIFIC INSTRUMENT MANUFAC-TURERS' ASSOCIATION have recently produced a second edition of their brochure on Radio Isotope Applications. The new edition summarizes all the equipment necessary for use with radio isotopes in all applications and gives the sources of such equipment in this country. Copies of the brochure are available free of charge on application to The Secretary, Scientific Instrument Manufacturers' Association, 20 Queen Anne Street, London, W.1.

THE RAGOSINE OIL COMPANY, LTD., have produced a leaftet giving details of their molybdenized lubricants. Their general uses, range and application are mentioned, together with grades and prices. The incorporation of molybdenum disulphide in a range of orthodox lubricants represents a definite step forward in the science of lubrication. Ragosine Oil Company, Ltd., Minerva Works, Woodlesford, Nr. Leeds.

or lubrication. Ragosine Oil Company, Ltd., Minerva Works, Woodlesford, Nr. Leeds. WIGGIN NICKEL ALLOYS--No. 18. The latest issue of this journal contains a number of interesting articles describing some industrial applications of various high-nickel alloys. Because of the diversity of uses of these materials, the articles cover a wide range of interests. Of special interest is an article on the Actarc Monta welding transformer which incorporates a Monel spindle in the variable current control. Other articles deal with the use of Monel staybolts in British Railways standard locomotives and with pickling crates included are detailed descriptions of the Bristol Proteus engine and of the Allen gas turbine with notes on their use of the various Nimonic alloys for turbine blades and combustion chambers. Copies may be obtained free from Henry Wiggin & Co., Ltd., Wiggin Street, Birmingham, 16.

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JUNE 1953

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G10/241E "Nomotron" Unidirectional Cold-Cathode Gas Tube Decade Counter

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49



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Maximum input voltage (r.m.s.)	125V	125V	125V	125V	250V
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ELECTRONIC ENGINEERING

JUNE 1953

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