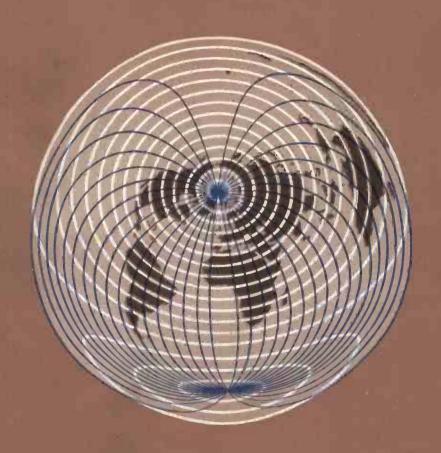
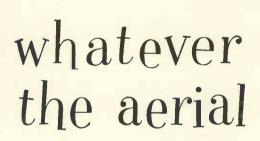
FEBRUARY 1957 TWO SHILLINGS

Wireless World

ELECTRONICS
Radio · Television



PORTY-SIXTH YEAR OF PUBLICATION

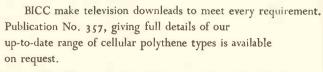


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Wireless World

ELECTRONICS, RADIO, TELEVISION

Managing Editor: HUGH S. POCOCK, M.I.E.E.

Editor:

H. F. SMITH

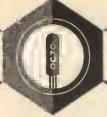
Assistant Editor; F. L. DEVEREUX, B.Sc.

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A Transistor



Controlled Regulated Power Unit

The circuit diagram of the regulated power unit, shown below, demonstrates a novel application of transistors as control devices. They are especially suited for this type of circuit because of the high voltage gains which can be obtained with them when they are operated with large collector loads from high voltages. In the power unit described, a Mullard OC73 is used to control the current flowing through an EL38 series valve. One advantage of using a transistor in this type of regulated power unit is that a separate negative h.t. supply is eliminated.

The power unit has an output impedance of less than three ohms, and will deliver 100mA at any voltage between 40 and 84 volts regardless of quite large changes in the mains input voltage. For instance; if the mains input voltage rises from 198V to 242V, the output voltage will increase by only 0.4V at full load. Reducing the load from 100mA to zero produces a rise in output voltage of only 0.3V at 220V mains input.

CIRCUIT OPERATION

By connecting the transistor as shown in the circuit diagram it compares the output voltage with the reference voltage, and any difference produces a large change in collector current due to its high effective mutual conductance.

Since a transistor can be operated at very low currents, very high voltage gain is obtained by connecting the collector to negative h.t. through a load of $500k\Omega$.

The collector is connected also to the control grid of the EL38 through a 100kΩ grid stopper. Therefore the collector-to-base voltage of the transistor is equal to the grid-to-cathode bias of the valve and is practically independent of the output voltage setting.

The output voltage is approximately equal to the reference voltage and can only differ from it by the baseto-emitter voltage of the transistor, which will be less than 0.1V. So the

to-emitter voltage of the transistor, which will be less than 0.1V. So the maximum output which can be obtained is determined only by the reference voltage. Although an 85V reference level is used in the circuit described, the design is almost identical for any reference level. The minimum output voltage of this supply unit cannot be set below the bias voltage of the series valve and therefore it is limited to about 40V.

When the power unit is working, a drop in output voltage will cause the base of the OC73 to become negative with respect to the emitter. As a result, the collector current increases and the collector voltage becomes more positive, thus reducing the bias on the EL38 and compensating for the original change. Equilibrium will be reached when the output voltage reaches the same value as the emitter voltage.

Variations in the mains voltage have only a slight effect on the reference voltage, which is stabilised by the 85A2, and therefore the output voltage remains substantially constant.

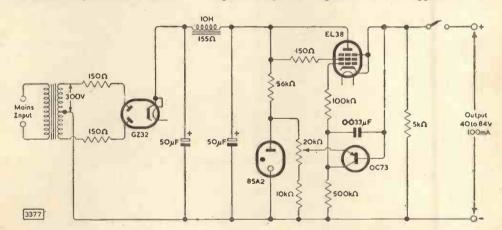
CIRCUIT DESIGN

The full-wave rectifier circuit is of conventional design, with choke smoothing to give a ripple-free output to the series valve, an EL38, and to the voltage reference tube, an 85A2. The 85A2 is operated at a low burning current to minimise the change in the reference voltage when the circuit is loaded and the poor regulation of the rectifying circuit causes the unstabilised voltage to fall.

An EL38 was chosen for the series valve because it has a comparatively short grid base. This is important since it governs the maximum collector-to-base voltage when no current is being taken from the supply. To prevent the collector voltage exceeding the - 30V collector-to-base rating of the OC73, the output must be permanently loaded to about 10mA; $5\,\mathrm{k}\,\Omega$ forms the permanent bleeder resistor.

The stability of the output is limited by the stability of the reference source, and by using an independent reference voltage or running the 85A2 from a 150B2, the stability could be increased, and the output impedance reduced to at least half the value given.

Changes in temperature have negligible effect on the output voltage and impedance but an appreciable in-



crease in ambient temperature raises the minimum transistor leakage current and slightly increases the minimum output voltage to which the supply can be set. This effect limits the maximum value of collector resistance which can be used. For operation at higher temperatures a smaller collector load resistance should be chosen.

Instability may occur when the power supply is first switched on if it is connected directly to a high current load, so a switch has been included in the circuit and should not be closed until the reference tube has ignited. The $0.033\mu F$ capacitor between the base and collector prevents high frequency oscillation which can occur under certain conditions.



Reprints of this series of advertisements WILL NOT be available

EDITORIAL COMMENT

GUARANTEES AND GOODWILL

Some time ago there was a newspaper story about a disgruntled owner of a television receiver. After suffering from a series of breakdowns and failures to get satisfactory repairs, he took the set back to the dealer's shop and hurled it through the plate-glass window. Happily, the state of affairs revealed in the subsequent court proceedings hardly reflects the typical public attitude towards radio dealers, or, for that matter, towards broadcast receiver manufacturers. All the same, it goes to show there is not so much goodwill as we would all like to see.

Part of this lack of goodwill must be attributed to the unsatisfactory nature of the guarantees given by makers of broadcast receivers. As our contributor "Diallist" points out, these are far too complicated and in some respects appear to the purchaser to be ungenerous, to say the least, particularly in regard to valves and c.r. tubes. It will be interesting to watch the effect on the public of the recent announcement by the makers of Ambassador and Baird television sets, who are now giving an extra year's guarantee with the c.r.ts.

During last autumn there was a somewhat bitter correspondence in *The Times* on the onerous nature of guarantees. Though some of the letters were directed against makers of "consumer durables" in general,

the radio industry got rather more than its share of reproach.

It may be argued that there is virtually no difference between the terms of the guarantees commonly given with broadcast sets and with other comparable articles. That may be true enough, but would Wireless World be excessively starry-eyed in thinking the time may well have come for the radio industry to lead the rest in trying to establish a brand-new kind of relationship between seller and buyer? The old legal maxim "let the buyer beware" seems to be utterly outmoded in a world where the complex products of modern technology go into the homes of buyers whose lack of knowledge leaves them completely in the hands of the sellers.

RADIO MILESTONES

Heinrich Hertz was born exactly one hundred years ago. Oversimplifying a little, it is often said that Marconi, by connecting an effective aerial to Hertz's oscillator, made the first significant step towards a workable radio system. It is true enough to say, in the same vein, that the second fundamental advance was made in 1906 when de Forest put a grid into Fleming's diode.

During the past year Dr. Lee de Forest has received many well-deserved honours to mark the jubilee of his invention. But the triode made virtually no impact on radio technique for at least six years. It was not until regeneration was introduced in 1912 that the triode offered outstanding advantages over existing devices. The regenerative receiver, with a sensitivity vastly greater than anything known before, opened up a new world, and also paved the way for the oscillating valve generator.

According to Armstrong, who opposed de Forest during prolonged patent litigation in America, the oscillating triode was the greatest of all radio inventions.

FEBRUARY 1957

Vol. 63 No. 2

High-Power Transistor Audio

DEMONSTRATIONS of audio amplifiers built round a new range of transistors were given recently by the General Electric Company. The demonstrations were preceded by some discussion of the preferred basic design features of the amplifiers.

Suitable collector bias conditions are chosen from the current/voltage characteristics and it is then thought best to measure the small signal a.c. relationships at these conditions in order to derive equivalent circuit parameters. By this method the internal feedback, which in transistors can be significant at all frequencies, can be more easily taken account of. This procedure is not as complicated as it may seem, for several of the equivalent circuit parameters are found to be roughly the same

in many circumstances.

Advantages of Class B Operation.—As regards the choice between class A or class B operation, there are stronger reasons for preferring class B with transistors than with valves. In transistors the collector dissipation, and the way in which this varies with the input signal, are both considerably different for the two types of operation. Thus if we take the maximum power limit as that determined by the maximum allowable mean collector dissipation, we find that the maximum available power for two transistors operating in class B push-pull is five times that for the same pair in class A; and is, in fact, five times the allowable collector dissipation. If we take the power limit as being that at which the distortion starts to increase rapidly (caused by alteration of current gain with peak current) similar ratios and powers are also obtained.

Another important consideration is the input power required. Owing to the characteristics of transistors they can be operated to much lower voltages relative to the h.t. supply than the corresponding pentode valves so that the maximum theoretical efficiencies for class A and class B operation of $\frac{1}{2}$ and $\frac{\pi}{4}$ respectively can be very nearly achieved using transistors. Under quiescent conditions the power drain with class B operation of transistors is of the order of 0.1 of the maximum output power rising to 1.3 times the maximum power when this is being delivered. With class A operation there is a continual drain of about four

times the maximum power.

Having regard to these two factors of available power and power input required, the best use of a.f. transistors will be made in class B battery amplifiers, i.e., in public address or in ordinary commercial battery receivers. Other factors supporting this view are their compactness and the low voltage supplies required. Furthermore, when a push-pull class B transistor amplifier is overloaded the "clipping" distortion produces little loss of clarity for public address purposes.

The common emitter type is preferred for both small and large signal amplifiers. This has the highest gain and any disadvantages can be conveniently avoided in circuits which reduce this gain. For example, the input resistance of such amplifiers is often undesirably low (e.g., when using a crystal pickup); and in order to increase this the simple

addition of a series resistor is most convenient in

spite of the loss of gain.

Phase Splitting Circuits.—In most of the amplifiers mentioned a transformer phase splitter was adopted. This gives a higher gain than if a transistor is used, but the size of the transformer may sometimes be a disadvantage. As the impedances concerned are low the transformer specification will be somewhat different from that in a valve amplifier. Thus the minimum primary inductance required for good l.f. response will be much less, being of the order of 50 millihenrys rather than 50 henrys. On the other hand the d.c. resistance must be very low necessitating thicker wire.

If a transistor is used to provide phase splitting the coupling condensers tend to charge up, and the forward bias on the output transistors necessary to reduce crossover distortion is decreased. The maximum available power is then decreased, though

the frequency response is improved.

In such transformer phase splitting circuits with transistors, several new types of distortion arise; but apart from the normal panacea of negative feedback other techniques are available for considerably reducing these. Thus the changing input resistance as one transistor takes over amplification from the other every half cycle (the emitter resistance varies inversely with the emitter current) can be avoided by a small amount of forward bias (≈100 mV) to the emitter-base junction so that a positive emitter current flows under quiescent conditions. Carrier storage effects producing ringing at a high frequency, which occur as well when the signal changes sign, can also be largely eliminated by bifilar winding of the secondary of the phase splitting transformer.

Phase changes in the transformers as well as in the transistors themselves of course limit the amount of feedback that can be supplied; but there is no difficulty in obtaining the roughly 7 dB overall feedback required to reduce distortion sufficiently in class B amplifiers, if an RC feedback loop is used to take some account of these phase shifts. The application of feedback in transistor circuits is also important for reducing the effect of variation among individual transistor characteristics in mass

production or replacement.

Complete Amplifiers.—Seven different amplifiers have been made with maximum powers ranging from 250 milliwatts to 20 watts. They consist of an input amplifying stage, the phase splitting transformer, and a push-pull output stage. For the class B amplifiers the total harmonic distortion at maximum power was ≈8 %. The 4-watt class A amplifier which was demonstrated (corresponding to the 20-watt class B amplifier using the same transistors) had 0.8 % total harmonic distortion at 4 watts (0.5 % at 2½ watts) with 16 dB overall feedback. This distortion could have been reduced to 0.3 % at 4 watts by using a class A push-pull driver stage. Power gains for both types of amplifier were ≈50 dB or greater.

The 250 mW amplifier corresponds to the usual battery radio valve output stage. However, in normal

Amplifiers

NEW RANGE OF TRANSISTORS

use the volume control is often set so as to cause overloading; so that an improvement in quality can be effected using a 1-watt amplifier, though a disadvantage is that the quiescent current drain is nearly doubled. There is no corresponding battery valve equivalent in this case.

A very compact battery-operated, 45 r.p.m., record player using the 1-watt amplifier was demonstrated. The battery life when supplying the turntable as well as the amplifier is about 15 hours using 5 U2 batteries. Although this may not seem very much; as the G.E.C. representative remarked, "15 hours worth of records take a very long time to play!"

The 250-mW amplifier for the output of a transistor a.m. receiver which was also shown in a very compact form, used a directly coupled 120-ohm centre-tapped loudspeaker voice coil. This increases the thermal stability in the output stage because of the d.c. resistance, but the application of overall feedback in two balanced loops is difficult. A singleended push-pull output stage was described as a means of rapidly changing to twice the power supply voltage for a normal push-pull output stage; but there was no direct coupled loudspeaker application for this case.

As regards other applications of these transistors, low-noise pre-amplifiers developed include the usual bass and treble tone controls; and for a normal microphone (600-ohm output impedance) or pickup ("variable reluctance" or crystal) input signal, give signal to noise ratios of about 60 dB. Concerning the possibility of a transistorized tape deck the most difficult problem is the provision of erase power. Two EW70s (one of the new types of transistor) can be used in class B push-pull to deliver 2 watts at 35 kc/s, thus largely solving this problem.

Details of New Range.-The new G.E.C. tran-

sistors are all of the germanium junction p-n-p type. They are hermetically sealed and of all-metal construction; the metal can provides a heat sink, and facilitates the attachment of a radiator. The lowresistance base connection, which is also integral to the construction, besides providing good thermal contact, also enables the potential h.f. performance to be more fully realized.

Having regard to their general characteristics and maximum allowable collector dissipation there are essentially three new transistor types. known as the GET 4 and 6, GET 5 (formerly EW70) and the EW57 (provisional); and have allowable collector dissipations (at 45°C) of 50 mW, 400 mW and 4 watts respectively. The frequency cut-offs are \approx 250 Kc/s for the EW57 and greater than 1 Mc/s

for the others.

The GET 6 is a low-noise version of the GET 4 and has a noise level well comparable with that in thermionic valves. The EW57 is also subdivided into three types according to the supply voltage required (6, 12 or 24 volts). All of these transistors have been in pilot production for about two years and should be in quantity production later this year.

With regard to the limiting operating conditions for these transistors there was thought to be good prospects of improvement. At present the maximum operating temperature is 50°C, but this can probably be increased up to about 70°C. If collector leakage currents become embarrassing silicon transistors (samples of which should be available before the end of the year) would be a complete answer and should function to well over 100°C

The power limit should also see a major improvement with operation at high collector temperatures. Moreover experimental samples which maintain their current gains at high values of emitter current (another limiting factor) have also been produced.

LITERATURE COMMERCIAL

Printed Circuits.—An illustrated booklet for engineers and technicians on the available types of printed circuits and the problems of designing equipment using them. Rigid, flexible and flush-bonded types are discussed, also incorporated components, heating elements, and facilities offered for development and production. From Technograph Printed Circuits, 32, Shaftesbury Avenue, London, W.1.

Tape-to-Disc Transfer Service.—Details and prices in a leaflet from Sound News Productions, 59, Bryanston Street, Marble Arch, London, W.1.

Sound Reproduction Equipment ("New Orthophonic High Fidelity"), including amplifiers, loudspeakers, tuner, transcription unit and pickups. Price list from R.C.A. Great Britain, Lincoln Way, Windmill Road, Sunbury-on-Thames, Middlesex.

High-Conductivity Copper Alloys, containing silver, cadmium, chromium, and tellurium. The last three decrease the conductivity slightly but give other desirable properties such as increased strength, resistance to wear and machinability. An informative booklet of 54 pages, containing many tables of properties, from the Copper Development Association, 55, South Audley Street, London, W.1.

response over 40-10,000 c/s. Leaflet from RGA Sound Services (Plymouth), 6, Conway Gardens, Enfield, Middlesex.

Radio and Electronics Books including the Philips series on valves, etc., listed in a catalogue from the Cleaver-Hume Press, 31, Wright's Lane, Kensington, London, W.8.

Small Capacitors suitable for loudspeaker crossover networks, claimed to occupy only 25-30% of space required by conventional types. Capacitances between 2 and 16 µF, working voltage 150 V d.c. Leaflet from A. H. Hunt (Capacitors), Wandsworth Road, London, S.W.18.

Variable Output Transformers (Regavolt) with self-aligning brushes to ensure maximum surface contact with winding. Four new types for normal mains voltages, with outputs variable over 0-275 V, currents between 6 and 10 A, and one new type for 115-V mains and output of 0-135 V at 15 A. Leaflets from the British Electric Resistance Company, Queensway, Enfield, Middlesex.

Microphones; crystal, ribbon, carbon and noise-cancelling types; also stands, table bases, transformers and other accessories. Illustrated catalogue from Lustraphone, St. George's Works, Regents Park Road, London, N.W.1.

Strain-Gauge Bridge, giving direct reading in percentage strain over the range 0.001% to 0.5%. Accuracy of measurement, $\pm 1\%$. Leaflet from the Croydon Precision Instrument Company, Hampton Road, West Croydon, Surrey.

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WORLD OF WIRELESS

I.S.M. Interference

RADIO interference from industrial, scientific and medical equipment is being considered by a committee set up by the Postmaster-General to advise him on the making of regulations prescribing limits of radiation.

The Wireless Telegraphy Act prescribes that the members of such advisory committees should either "possess expert knowledge of the matters falling to be dealt with" or "represent persons whose interests are likely to be affected" by any regulations made. The nineteen-man committee, of which O. W. Humphreys is chairman, therefore covers the interests of those concerned with both the cause and effect and includes representatives of the B.B.C., I.T.A., equipment manufacturers, air navigational specialists, production engineers, the medical profession and the viewer and listener.

V.H.F. Coverage

BY the opening of three more v.h.f. broadcasting stations (Wenvoe, replacing the temporary low-power transmitter, Sutton Coldfield and Norwich) a day or two before Christmas, the B.B.C. made good its promise to complete the first batch of ten stations by the end of 1956. To say "complete" is perhaps a slight exaggeration, for the Cardiganshire station at Blaen Plwy at present has only one of its three transmitters (Home Service) working. An eleventh station, at Penmon, was subsequently added to the original chain, but so far only one transmitter has been installed and this is radiating the Home Service with an e.r.p. of 1 kW.

The service now reaches 84% of the population.

N. Hessary Tor (S. Devon) Sutton Coldfield (Warwicks.) Pontop Pike (Co. Durham)	Light (Mc/s) 88.1 88.3 88.5	Third (Mc/s) 90.3 90.5 90.7	Home (Mc/s) 92.5 92.7 92.9	e.r.p. (kW) 60 120 60
Meldrum (Aterdeenshire) Blaen Plwy (Cardigan.) Wrotham (Kent) Holme Moss (Yorks.) Penmon (Anglesey) Norwich Wenvoe (Glam.) Divis (N. Ireland)	88.7 88.7 89.1 89.3 89.6 89.7 89.9 90.1	90.9 90.9 91.3 91.5 91.8 91.9 92.1 92.3	93.1 93.1 93.5 93.7 94.0 94.1 94.3 94.5	60 60 120 120 60 120 120

U.K. Display in New York

THE Institute of Radio Engineers' annual convention and show, which last year attracted 714 exhibitors, is being held in New York from March 18th to 21st. The Institute has offered an area of 1,200 sq ft for a collective U.K. display of radio and electronic equipment. Manufacturers interested in participating should communicate at once with the Board of Trade, Exhibitions and Fairs Branch, Lacon House, Theobalds Road, London, W.C.1. (Tel.: Chancery 4411, Extn. 436.)

Organizational, Personal and Industrial Notes and News

Tape Recording Patents

THE Armour Research Foundation has carried out extensive research during the past 15 years on all aspects of magnetic recording, and has filed over one hundred American patents; most of these have also been granted in the United Kingdom.

John P. Skinner, manager of the A.R.F. Development Corporation, a subsidiary of the non-profit-making Foundation, stated during a recent visit to this country that to make an effective tape deck it is necessary to use principles contained in at least one of the A.R.F. patents.

It is understood on enquiry that licensing agreements have already been concluded by the A.R.F. Development Corporation, of Chicago, Illinois, with the following companies in this country: Boosey and Hawkes, Collaro, Garrard, E.M.I., Grundig, Simon, Tape Recorders (Electronics), Verdik Sales, Walter (Tape Recorders), and Wright and Weaire.

Amateurs and TV Interference

A NEW policy regarding amateur interference with sound and television reception has been announced by the Post Office. In the past, if complaints were received of an amateur causing interference to television reception due to "blocking," the Post Office prescribed that he must not transmit during television broadcasting hours.

Under the new arrangement if an amateur is otherwise transmitting within the terms of his licence, but causes interference to sound or television reception and it can be demonstrated that a reasonable remedy, such as the fitting of a simple filter, is available to the owner of the receiver, "then the amateur will be allowed to continue operating after an interval of one month from the time at which the cure is explained and demonstrated to the complainant by the Post Office."

It is understood that this will apply to all cases of interference to sound and television reception where the amateur's transmissions are found to be within the terms of his licence.

Although B.R.E.M.A. has been informed of the new policy no official announcement has so far been made on behalf of receiver manufacturers.

NEW YEAR HONOURS

A number of those who played a leading part in the planning, production and laying of the Atlancic telephone cable were recipients of awards in the New Year's Honours List. They include J. N. Dean, chairman of the Telegraph Construction and Maintenance Company (Knighthood); R. J. Halsey, an assistant engineer-inchief at the G.P.O. (C.M.G.); A. H. Roche, telecommunication engineer in charge of Submarine Cable System Development and Production Division, Standard Telephones & Cables (O.B.E.); E. F. Neve, foreman S.T.C. submerged repeater manufacturing shop (B.E.M.); and E. V. T. Perrins, technical officer Post

Office Research Station, who has specialized in repeater test equipment (B.E.M.).

Sir Stanley Angwin, who recently retired from the chairmanship of the Commonwealth Telecommunication Board, and E. M. Jones, director, the Government Communications Headquarters, Foreign Office, are appointed K.C.M.G.

Three members of the G.E.C. Research Laboratories staff received awards:—O. W. Humphreys, director, is appointed C.B.E.; E. G. James, head of crystal valve development, O.B.E.; and W. C. Cropper, group leader in charge of a special project for the Admiralty, M.B.E.

Among those appointed M.B.E. are Miss B. K. Chaplin, executive officer at the D.S.I.R. Radio Research Station; W. H. Hopkins, works manager, E.M.I. Factories; G. W. Kilm'ster, first radio officer, R.M.S. Arundel Castle; E. L. Lycett, assistant, B.B.C. outside broadcasts; H. Starr, signals officer, Civil Aviation Telecommunication Directorate, Ministry of Transport and Civil Aviation; and W. W. Syrett, export manager of Ekco's Radio Division.

Among overseas radio personalities who received honours are L. A. G. Hooke, managing director of Amalgamated Wireless (Australasia), a knighthood for "services to the radio industry in Australia"; T. W. Chalmers, formerly director of broadcasting in Nigeria, C.B.E.; T. V. Hooley, officer-in-charge, Government Wireless Station, Falkland Islands, M.B.E.; and E. R. Richardson, senior assistant controller of telecommunications, Nigeria, M.B.E.

PERSONALITIES

Sir Noel Ashbridge, who retired from the directorship of technical services of the B.B.C. in 1952 and became a director of Marconi's, has been elected an honorary member of the Institution of Electrical Engineers. The election is in recognition of his services to the Institution, of which he was president in 1941, and of "his outstanding contributions in the field of radio engineering, particularly in the development of the British sound and television broadcasting services."

After what must be a record for service with one radio company—over 54 years—R. D. Bangay has retired from Marconi's. In 1952 he celebrated his jubilee in the radio industry, the first man to do so. Soon after joining the company in 1902 he went to America for five years and helped in the installation of the first U.S. coastal radio station, at Babylon. He was concerned with the original experiments in air-to-ground communication, and in 1914 was placed in charge of the company's department established for the development of military radio equipment. Mr. Bangay was for some years chief of designs, and since 1935 has been foreign manager. He was the author of two books published by Wireless World many years ago: "The Elementary Principles of Wireless Telegraphy" and "The Oscillation Valve."

L. C. Jesty, B.Sc., M.I.E.E., has left Marconi's, where for seven years he led the television research group, and has joined the Sylvania-Thorn Colour Television Laboratories at Enfield, Middlesex. He will be in charge of colour television research. Whilst at Marconi's research laboratories he had been closely concerned with the Anglicized version of the American colour system which has been used experimentally by the B.B.C. From 1927 to 1946 Mr. Jesty was on the staff of the G.E.C. Research Laboratories where in 1933 he started the c.r.t. research group which was responsible for the introduction of the VCR97. For three years prior to joining Marconi's he was in charge of advanced research at the Cinema-Television Laboratories.



The Radio Communication and Electronic Engineering Association has created the new post of technical secretary, to which J. F. Richardson, Grad.I.E.E., has been appointed. He will work under H. E. F. Taylor, recently appointed executive secretary. Mr. Richardson began his engineering training in 1939 with S.T.C., and since 1952 has been technical assistant to the director, Electric Lamp Manufacturers' Association.

Reginald A. Yeo has left the Admiralty Signal and Radar Establishment, at Haslemere, where he was head of the electronics division, on being appointed full-time member of the Australian Broadcasting Control Board. He was a principal scientific officer in the Royal Navy Scientific Service, and was concerned mainly with radio matters throughout his government service. He was delegate at the conference of the International Telecommunication Union at Atlantic City in 1947.

G. B. Jeffery, M.A., B.Sc., A.M.I.E.E., has left the Royal Aircraft Establishment, Farnborough, where for four and a half years he has been senior engineer in the data transmission section of the Electrical Engineering Department, and has joined R. B. Pullin and Co. He is technical sales manager of the Pullin-Kearfott Division, which, under a recent licensing agreement with the American Kearfott Company, Inc., will manufacture synchros, servomotors and tachometer generators. During the war Mr. Jeffery was a radar officer, R.N.V.R., and was on the staff of H.M. Radar School, H.M.S. Collingwood.

H. L. A. Foy, the new publicity manager of Decca Radar, Ltd., has been engaged on the operational aspects of radar since he joined the company two years ago. He was a specialist navigating officer in the Royal Navy, and since joining Decca has been particularly concerned with the introduction of "True Motion" radar.

C. Hardy and C. B. Speedy, Ph.D., B.E., Grad.I.E.E., have been appointed directors of Data Recording Instrument Company, Ltd., of Feltham, Middlesex, which previously operated as a division of Lion Electronic Developments, Ltd. Mr. Hardy was for many years at the Signals Research and Development Establishment of the Ministry of Supply, where he was working on data recording. Dr. Speedy has been engaged since 1945 on the study of electronics especially in relation to computers.

J. N. Macleod and C. Metcalfe have been appointed directors of Electric and Musical Industries, Ltd. Together with E. J. Emery, who was appointed a director last February, they have been appointed managing directors of the company, responsible, respectively, for the U.K. Record Division and the Oversea subsidiary companies (other than Capitol Records Inc. and its subsidiaries), the E.M.I. Electronics Division (commercial and industrial equipment), and the Domestic Electronics Division.

H. A. Lewis, M.B.E., T.D., B.Sc.(Eng.), A.C.G.I., M.I.E.E., who recently left Marconi's to become personal assistant to E. J. Emery, managing director of the Domestic Electronics Division of E.M.I., has been appointed a director of E.M.I. Sales and Service, Ltd.

D. L. Johnston, who in our October, 1954, issue described a transistor replacement unit for hearing-aid h.t. batteries, has left Fortiphone, Ltd., where he was manager of the component division, and has joined Automation Consultants and Associates, Ltd.

The following appointments are announced by Belling and Lee: D. W. Rippin, who has been outside technical representative, becomes export manager; J. E. Bailey, B.Sc., and R. E. Meldrum join the company as technical representatives to manufacturers, and A. Fender is appointed technical sales representative to the trade in the area centred on Newcastle-upon-Tyne.

L. C. Smith has resigned from Plessey's which he joined ten years ago as technical representative. During the last war he was lieutenant-commander in the electrical branch of the Royal Navy and specialized in the maintenance of asdic. Before the war he ran his own service business in Birmingham for five years, prior to which he was a service technician with E.M.I. His address is 122, Whitaker Road, Derby.

Peter E. M. Sharp, A.C.G.I., B.Sc.(Eng.), A.M.I.E.E., whose three-year contract with China Engineers, Ltd., has ended, has returned to this country. He joined the Telegraph Construction and Maintenance Company in 1951, subsequently transferring to their agents in the Far East. His home address is 46, Hyde Vale, London, S.E.10.

OUR AUTHORS

Dr. E. L. C. White, of the research division of E.M.I. Electronics, who recently addressed members of the Television Society on alternatives to the American television system, contributes an article on page 75 covering some of the points discussed. Dr. White has been with E.M.I. since 1933 and was closely associated with the late A. D. Blumlein in developing the Marconi-E.M.I. television system adopted by the B.B.C. During, and to some extent since, the war, he has been concerned with the development of radar display systems, but has more recently worked on colour television. Before joining E.M.I. he was for three years at Cavendish Laboratory, Cambridge, working on pulse methods of ionospheric research. He is 47.

Michael P. Beddoes, contributor of the article on an improved sync separator, went to Canada a few months ago, and is now assistant professor in electrical engineering at the University of British Columbia. After graduating in electrical engineering at Glasgow University he spent seven years in industry, and then in 1953 went to the City and Guilds College where he worked on television band-compressing systems. His industrial experience started at G.E.C., Coventry (as a post-graduate apprentice), and after service as a senior engineer at Plessey's he became assistant chief engineer of McMichael's radio division,

IN BRIEF

During November the number of television licences increased by 142,345 bringing the total to 6,433,417. **Broadcast receiving licences,** including those for television and 310,690 for car radio, totalled 14,424,236 at the end of November.

Brit.I.R.E. Constitution.—An extraordinary general meeting of members of the British Institution of Radio Engineers is being held on January 30th, at which it is proposed to redraft the constitution. It is understood that the whole of the articles of association are being revised to give emphasis to the Institution's coverage of electronics.

"Wireless World" Index.—The publication of the Index for 1956 has been unavoidably delayed; it will be available towards the end of February. Cloth binding cases with index cost 7s 6d (postage 6d). Our Publishers will undertake the binding of readers' own issues, the cost per volume, including the index and binding case, being 22s 6d, plus 1s 6d postage on the bound volume.

Bilingual Television.—The system for transmitting two sound programmes in a single sound channel described on p. 79 has actually been developed for use by the French television service in Algeria, which has both French and Arabic viewers. A complete description, with a photograph of the "decoder" unit, appears in the January issue of the French journal Télévision.

American TV in Germany.—Television on American standards is being radiated from two transmitters in Germany for the benefit of U.S. service men. The low-power stations, which operate in Band IV (470-585 Mc/s), are at Bitburg and Landstuhl. They transmit films, provided by the Armed Forces Television centre at Limestone, Maine, for seven hours a day.

The French Components Show will be held from March 29th to April 2nd at the Parc des Expositions, Porte de Versailles, Paris.

America's Institute of High-Fidelity Manufacturers is sponsoring a **Hi-Fi Show** in Los Angeles from February 6th to 9th.

An electronic computing service for industrialists and scientists has been introduced by the Battelle Institute at Frankfurt/Main, Germany. Enquiries regarding the service and the facilities provided are being handled in this country by the Electronics Division of Remington Rand, Ltd., Commonwealth House, 1-19, New Oxford Street, London, W.C.1.

Interference Suppression.—Another conference on radio interference (the third) is being organized by the Armour Research Foundation of Illinois Institute of Technology. It will be held in Chicago on February 26th and 27th.

Colour Television.—Sales of R.C.A. colour television receivers were expected to reach the forecast 200,000 for 1956. It is anticipated that half a million colour receivers will be produced by R.C.A. this year.

Special courses in higher technology, being held during the spring and summer terms at colleges in London and the home counties, are listed in a bulletin issued by the Regional Advisory Council for Higher Technological Education. It costs 1s 6d and covers a wide variety of courses, including radio and allied subjects, being held at twenty-nine colleges.

I.o.M. Television.—A permanent television station, to replace the temporary one which has been in use since December 1953, is to be built by the B.B.C. at Carnane, near Douglas, Isle of Man, by the end of this year. The original plan to build the permanent station on Snaefell has been dropped.

Peterborough is having a three-day audio fair starting on January 22nd. It is being organized by Camera and Cine Centre, of 14, Long Causeway, at the Grand Hotel where lectures and demonstrations are being given by representatives of audio equipment manufacturers.

Each Saturday from 10 to 12 noon demonstrations of audio equipment are being given by Pamphonic Reproducers at their showrooms at 17, Stratton Street, London, W.1.

An inexpensive (25s) single-stage transistor audicamplifier, intended for working with a crystal receiver is being made by Warren's Radio, 88, Wellington Street Luton, Beds.

A bibliography of high-fidelity sound reproduction has been compiled by K. J. Spencer; and is available from the Library Association, Chaucer House, Malet Place, London, W.C.1; price 2s 6d. Approximately three hundred references, mostly later than 1947, are given. A more extensive bibliography of the subject is being prepared.

"Portable Transistor Superhet."—The germanium diode detector in this circuit (January issue) has been drawn with the wrong polarity of connections. It should be reversed, so that a positive-going a.g.c. voltage is applied to the base of the 1st i.f. transistor.

R.S.G.B. Membership.—The annual report of the Radio Society of Great Britain records that whereas during the past five years membership had declined by nearly 5,000, last year's decrease was only 57. The number of licensed amateurs in the Society actually increased by 95, bringing the total to 5,141. Non-transmitting members total 2,961.

Winning Design in the competition sponsored by the British Plastics Federation "to encourage young craftsmen to design articles in plastics materials" was for a 17-in portable television cabinet. The designer, E. J. Arundell, of Liverpool, receives 50 guineas.

Two 16mm Mullard sound films—one on cathode-ray tubes (lasting 32 minutes) and another on quality valves (27 minutes) are now available on free loan from the Central Film Library, Central Office of Information, Government Building, Bromyard Avenue, Acton, London, W.3.

The products and services of over seven thousand member firms of the Federation of British Industries, listed alphabetically under more than 5,400 headings, as well as lists of trade marks and trade names, are included in the 1957 edition of the "F.B.I. Register of British Manufacturers." French, German and Spanish glossaries are also incorporated in the Register, which is obtainable from our publishers, price two guineas, post free.

E.I.B.A.—A number of radio manufacturers are listed as donors to the Electrical Industries Benevolent Association in its 1956 Year Book. In addition, the Radio Industry Council gave £500, the B.B.C. £158, and the Radio Industry Clubs of London and Glasgow £350 and £140, respectively. The object of the Association is to assist deserving and necessitous persons who are, or have been, in "any branch of an electrical industry."

BUSINESS NOTES

Nine film scanners, each employing two film projectors and two television cameras, have been ordered by the B.B.C. from Pye. A feature of the equipment is that by the use of movable mirrors either camera can be focussed on a projector, thus minimizing the possibility of breaks in transmission.

An underwater television camera, which can be held by a frogman or diver, towed or fitted to a vessel, has been produced by **Pye**. Spherical in shape, it is intended for use at depths down to 3,000 feet. It incorporates a depth indicator from which readings are conveyed to observers on the surface.

Type approval, covering humidity and temperature, has been granted by the Joint Service Radio Components Standardization Committee for Ferranti's 2-in sealed instruments—voltmeters, ammeters and milliammeters. Their $2\frac{1}{2}$ -in and $3\frac{1}{2}$ -in instruments are classified as "design approved."

Communication on six v.h.f. channels is provided in the air traffic control system recently installed by International Aeradio at Vickers-Armstrong's airfield at Wisley, Surrey.

Two limiter stages are incorporated in the new Orthophonic high-fidelity f.m. tuner being produced by RCA Great Britain. The circuit employs seven valves, two crystal diodes and a tuner indicator.

Sales abroad accounted for £33M of the E.M.I. group's £53M turnover during the year ended June, 1956. The year's total was an increase of £11.4M on the previous twelve months.

Orders for millimetre-band telecommunication test equipment valued at over £25,000 have been placed with Marconi Instruments by the Ministry of Supply. Each 6-ft rack-mounted assembly comprises electronically regulated power supply and a frequency stabilization system.

A distribution centre, including showrooms and service information department, for Ambassador and Baird receivers, has been opened at 131, Great Suffolk Street, London, S.E.1 (Tel.: Hop 0791). K. H. Yandell, sales director of Ambassador Radio and Television, Ltd., and F. Duer, southern area sales manager, are now at this address.

Shirley Laboratories, Ltd., which was formed in 1954 by A. W. Wayne (a contributor to Wireless World) for the manufacture of amplifiers, tuners and electronic instruments, has moved to 3, Prospect Place, Worthing, Sussex. (Tel.: Worthing 30536.)

Tape Recorders (Electronics), Ltd., have recently started production at their second factory at 784-788, High Road, Tottenham, London, N.17. The price of their Sound Cadet recorder has been reduced to 39 gns.

S.S. Electronics, Ltd., recently moved from Harrow to Chiltern Works, Severalls Avenue, Chesham, Bucks. (Tel.: Chesham 8909.)

The midland office of Marconi Instruments, Ltd., formerly situated at 19, are now at 24, The Parade, Leamington Spa.

Goodmans loudspeaker cabinet, formerly known as the Viscount and recently renamed Canberra, is now called Sherwood.

OVERSEAS TRADE

Since the formation of Decca Radar og Navigator A/S, Bergen, the Norwegian subsidiary of the Decca Radar and Navigator Companies, in February, 1955, orders have been secured for radar installations in more than 300 Norwegian ships. It has also established a chain of service depots. The company's new general manager is F. I. Willoch. He has taken over from E. Tyler who has returned to the London office.

According to a recently completed analysis of places to which Britain exports radio and electronic gear, the United States was the largest buyer of British radio equipment in the first six months of last year. The value of U.K. radio exports to the U.S.A. was £1.6M, nearly 8 per cent of the industry's total overseas trade. Incidentally, the bulk of this was for sound reproducing equipment.

Marconi's are installing a temporary 2-kW medium-wave broadcasting transmitter at Brunei, on the north-western coast of Borneo, preparatory to setting up a permanent 20-kW transmitter on Tutong, some 35 miles from the town. The permanent transmitter will be fed by a frequency-modulated radio link from Brunei. A receiving station with double reversible rhombic aerials in dual diversity for the reception of B.B.C. and Australian stations, and dipole arrays for the reception of less distant stations, is also being built.

A contract for a v.h.f. multi-channel radio-telephone system for India's Western Railways has been placed with Marconi's. The radio-telephone links, which will have a capacity of 48 two-way channels, will be between Bhavnagar and Surat, and Jamnagar and Raikot.

Pye "Ranger" v.h.f. radio-telephone equipment, employing channel spacing of only 15 kc/s, was recently demonstrated in Toronto. This equipment is being made available immediately for those countries where there is intense frequency congestion.

Kelvin-Hughes echo sounders and radar have been fitted in the 47,000-ton tanker Evgenia Niarchos, the biggest yet built in this country.

Representation of U.K. manufacturers of loud-speakers, selenium rectifiers and tape recorders is being sought by John R. Tilton, Ltd., of 51, McCormack Avenue, Toronto 9, to whom illustrated literature, with ex-works and c.i.f. Canadian port prices in Canadian dollars, should be sent.

Output Transformerless Amplifiers a general review

HOSE of us who are in search of perfection in audio amplifiers must often have looked askance at the output transformer. In general it would be possible to reduce the non-linearities in an amplifier below any desired value by the application of a sufficient amount of feedback. However, as is well known, the output transformer produces undesirable phase shifts at the extreme ends of the audio spectrum which limit the amount of feedback which can be applied before instability sets in. These phase shifts also decrease the effective feedback at these extreme frequencies and this causes increased distortion in these regions. Varying core losses, hysteresis effects, matching variations, and incomplete coupling between the primaries also more directly increase the distortion, and this increase also is more pronounced at the frequency extremes.

Modern transformer design techniques of sectionalized windings, and particularly the use of C-cores, have to a large extent overcome these disadvantages in practice; but this has naturally given rise to an increase in price, and the fundamental

limitations still remain.

Increased Possibility of Class B Operation.— Using loudspeakers of normal efficiency and impedance (15-ohm) in an average living room an accepted peak power requirement from the amplifier is of the order of fifteen watts. In this case if there is no output transformer we will obviously require currents of the order of one ampere from

the output valves.

The problem of obtaining these currents is somewhat eased because in transformerless amplifiers class B operation of the output valves becomes more feasible than such operation is in an amplifier with an output transformer. Normally in class B operation using an output transformer incomplete coupling between the two half primaries produces transients when the valves cut off. These transients produce distressing audible distortion and are very difficult to eradicate, though a special bifilar transformer with both cathode and anode feeds designed by McIntosh and Gow¹ does succeed in doing so.

In class B operation owing simply to the higher outputs the general distortion is higher but this is not seriously so and can, of course, be reduced by

increased feedback.

High-Impedance Loudspeakers?—Allied to the difficulty of obtaining sufficient current from the output valves is that of matching, without too much distortion, the comparatively high valve impedance (of the order of a few thousand ohms) to the low impedance of the loudspeaker voice coil.

An obvious solution to both these problems is to use loudspeakers of higher impedance, but here the necessity of using thinner wire to keep the voice coil weight down produces its own problems. A few speakers of impedances in the range of 200-500 ohms have, however, been marketed, but there is not the usual variety of models available. Readers will realize the possibility that the newer electrostatic speakers will fit more smoothly into such a system, though their capacitive nature increases the matching problem.

In general by suitably paralleling output valves a transformerless amplifier for high-impedance loudspeakers can be adapted for low-impedance ones, so that the general features of the various designs can be considered without regard to the voice-coil im-

pedance.

Straightforward Methods.—A first approach is simply to use normal circuits with the loudspeaker directly replacing the output transformer or load, bearing in mind that it will often be more convenient to connect the speaker in the cathode or

low-voltage side of the valve.

Such designs generally involve capacitive coupling to the loudspeaker or direct current through the voice coil. Capacitive coupling requires almost impossibly high values of condenser for good low-frequency response and small phase shift in the case of low impedance loudspeakers, and may produce distortion due to hysteresis in the condenser.

Direct current in the voice coil would move it toward the positions of non-linearity for the suspension and non-uniformity of the magnetic field and in practice give a considerable increase in distortion. The possibility of increasing the linearity of the suspension (and thus decreasing the distortion), while maintaining sufficient restoring force to allow the required audio power to be developed, has already been largely exhausted in the design of

conventional loudspeakers.

Distortion caused by a non-uniform magnetic field could be avoided by making the voice-coil longer than the field so that the same length is always immersed in the field. This arrangement is often used in bass loudspeakers where the increased weight of the voice-coil is less important. An equivalent method would be to lengthen the magnetic field but the larger magnets required would increase the cost. The alternative of having the voice-coil initially asymmetrical with respect to the magnetic field would require a change in the initial displacement for different amplifiers and is thus rather impracticable.³

This simple approach also unfortunately needs a centre-tapped voice coil for normal push-pull operation. For "ultra-linear" operation two more tappings are even necessary! From the point of view of requiring as little and as practical a change as possible in existing loudspeaker design this approach

is seen to be inadequate.

Some sort of balanced arrangement whereby direct current in the voice coil is avoided would be an improvement. A straightforward circuit along

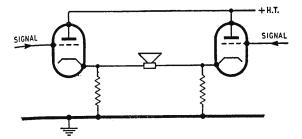


Fig. 1. Fletcher-Cooke output stage.

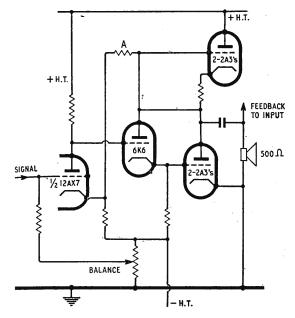


Fig. 2. Stephens amplifier.

this line is described by Fletcher and Cooke.⁴ The authors use a simple push-pull cathode follower arrangement shown schematically in Fig. 1. In order to obtain 12 watts of peak power in a 16-ohm loudspeaker 16 6AS7Gs were necessary. No distortion figures were quoted, and the damping was not very good (the output impedance was 23 ohms) but the authors did not use any overall negative feedback. In such cathode follower circuits low-impedance loudspeakers have an advantage in that the voltage requirements from the driver are not so serious as is usually the case.

Series Connected Output Stages.—Most other circuits use a series connected output stage, with either a single or push-pull input to this stage. In either case since the valve outputs in the load add together the optimum value for this load is less than half of that in the conventional push-pull arrangement. This type of circuit thus considerably eases the matching problem. It is also often easier to arrange d.c. connections between earlier stages and the output valves which improves the low frequency response and decreases the phase shift in this region.

Using a single input, series connected, output stage two commercial amplifiers have been developed, by Stephens⁵ and Philips,⁶ both for capacity connected high-impedance loudspeakers. Schematic

diagrams for these are shown in Figs. 2 and 3. In both cases the signal from the lower output valve is fed to the grid of the upper so that the input signal varies the voltage across the output valves in opposite directions. More complicated power supply arrangements could have avoided the necessity of capacity connections to the loudspeakers. circuit of the Stephens amplifier shows how d.c. connections between early stages may be arranged in this type of circuit. The resistor A provides negative feedback and serves to stabilize the bias on the 2A3s. With about 40 dB of feedback 3rd harmonic distortion is of the order of 0.4 per cent for 20 watts r.m.s. output into the 500-ohm load. Unfortunately, in both these amplifiers the voice coil does not provide a sole common load for the two output valves because they also partially load each other. Although cancelling of even harmonic distortion products could still be obtained by arranging for the output valves to give equal amounts of such distortion in the voice coil, this balancing would be difficult if not impossible, with the few variables

Push-Pull Series Connected Output Stages.—The last-mentioned disadvantage does not apply if the

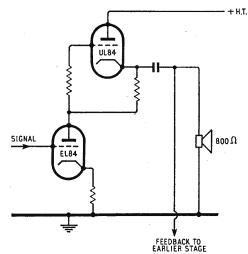


Fig. 3. Philips output stage.

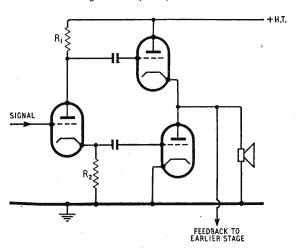
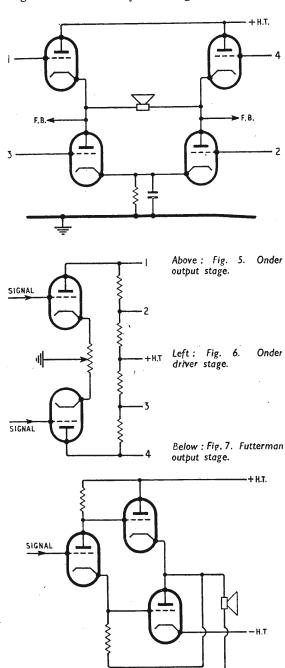


Fig. 4. Dickie-Macovski output stage.

output valves are series connected with push-pull input. In this case the optimum output impedance is one quarter of that for normal push-pull operation. If, however, the phase splitter and output valves are not correctly designed together it is difficult to retain equal drive in the output valves as the loudspeaker impedance changes.

An example of this difficulty arises in an amplifier described by Dickie and Macovski. Omitting a push-pull driver stage the phase splitter and output stages are schematically as in Fig. 4. Here it will



be seen that the voice-coil load is coupled so that it provides feedback to the grid of the upper valve. Thus the input to this valve has to be increased and the resistor R₁ is made greater than R₂. However the variation of voice-coil impedance over the frequency range (which may easily be of the order of 5 to 1) prevents exact adjustment by this means except over a narrow frequency band. In this amplifier the effect of this impedance variation is reduced by suitably shunting the voice coil; an example of such a shunt being simply a .01-µF condenser and a 16-ohm resistor in series. This prevents the rise in impedance at high frequencies produced by the inductive voice coil and thus avoids instability caused by increased feedback and phase shift. The anode loads of the push-pull driver stage could also be adjusted to give better balance.

In this amplifier there is one voltage amplifying stage before the phase splitter and 3 6082 valves (26.5 volt versions of the 6AS7G) operating nearly in class B in the output stage. With 40 dB of overall feedback the harmonic distortion was 0.4 per cent for 25

watts r.m.s. into a 16-ohm load.

Essentially the same balancing difficulty arises in a variation of the circuit of Fig. 4 described by Onder.8 Here instead of simply paralleling output valves to give increased power they are arranged in a bridge circuit (shown schematically in Fig. 5) which increases the optimum load to four times that for a parallel arrangement. Diagonal valves are run in phase, suitable driving voltages being obtained from a push-pull stage as in Fig. 6, this arrangement giving the necessary greater input to the upper tubes. An ordinary "concertina" phase splitter with suitable tappings on the loads could, of course, also be used. Here again using a balancing potentiometer as in Fig. 6, d.c. connections become possible.

In this amplifier there is a "see-saw" type of phase inverter before the driver, and 2 6AS7G valves operating in class A in the output stage. With about 15 dB overall feedback the intermodulation distortion was 0.7 per cent for 9 watts into a 400-ohm load.

Equalization of Drive in the Output Valves.—If we return now to Fig. 4 equal drive in the output valves can be obtained very simply as described by Futterman, by returning the earthy end of R₂ to the junction of the output valves as in Fig. 7. In this case both output valves are acting as cathode followers so that the voltage requirements from the phase splitter are large. However, the load is in the input to this valve in the correct sense to provide positive feedback so that a much lower voltage is actually required.

In the amplifier described in this reference (Ref. 9) there is a pentode, high gain, voltage amplification stage before the phase splitter. The cathode return from this pentode is taken to the tap of a potentiometer across the load so that varying amounts of negative feedback may be applied. There was in fact some difficulty in obtaining sufficient gain to give enough feedback to reduce the distortion sufficiently, but a phase splitter giving gain could be used. Using 14 type 12B4 valves operating in class B in the output stage, with 48 dB of overall feedback the harmonic distortion was 0.1 per cent for 20 watts into a 16-ohm load. Square wave tests on this and the amplifier of reference 7 already described give very impressive results even at

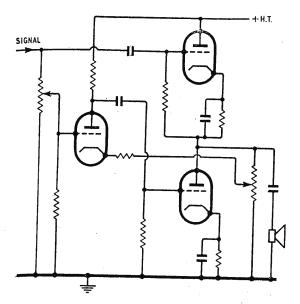


Fig. 8. Coulter output stage.

20 kc/s, with very short rise times and no overshoot. A more complicated way of obtaining this equal balance, by applying extra feedback to the lower output valve in Fig. 4 is described by Coulter¹⁰ and illustrated in Fig. 8. Here the upper output valve is fed directly from the input. A suitable fraction of this input has its phase inverted by another valve and is then applied to the grid of the lower output valve. By taking the cathode feed for the phase inverter from the tap of a potentiometer across the load, negative feedback is applied to this inverter and thus to the lower output valve. The amount of this negative feedback is adjusted to compensate for that produced by the load in the upper valve.

A disadvantage of this arrangement is that the distortion in the input to the lower valve is greater than that to the upper as this input has to pass through an extra voltage amplifying stage (the phase inverter); and the even harmonics of this extra distortion cannot be cancelled in the output stage. Furthermore, the degeneration caused by the speaker varies with its impedance, whereas the feedback is a constant fraction of the output voltage.

This general type of output stage has been recently applied to a transistor circuit by Cossor's. Transistors have an encouraging future in transformerless amplifiers as their voltage requirements are much lower, and a considerably higher overall efficiency should be possible. In the Cossor circuit equal drive in the output transistors is obtained by an input transformer. This transformer also avoids d.c. unbalance in these transistors but is contrary to the general philosophy of such amplifiers.

Returning again to Fig. 4, perhaps the best way of obtaining balanced inputs to the output valves is, as described by Peterson and Sinclair,¹² to take the anode supply for the phase-splitter from the junction of the output valves as in Fig. 9. In this case the input voltages to the output valves are always developed between cathode and grid, so that the load does not produce a cathode follower effect

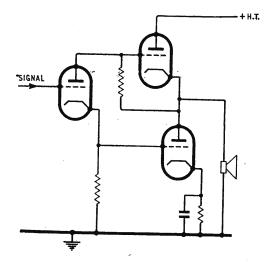


Fig. 9. Peterson-Sinclair output stage.

in either valve and there is no unbalance. A disadvantage is the negative feedback produced by the load on the supply voltage for the phase-splitter. This circuit also is susceptible to d.c. connection.

The authors give a general discussion of this type of amplifier using transformers solely as matching devices. No practical details of a strict transformer-

less amplifier are given.

Extended Class A?—Another novel type of circuit that may perhaps be of value in these amplifiers was given the name, "extended class-A," by the author.13 Here a triode and tetrode are run in parallel with their grids and anodes directly con-The valves are biased for normal class A operation for the triode and this usually cuts off the tetrode, so that the arrangement acts as a triode for small signals. When the signal becomes sufficiently large (usually about one third of the maximum) the tetrode starts drawing current and increasingly controls the operation. This gives the transfer characteristic a slight curve but this is not serious. The circuit should combine the advantages of the low output impedance of triodes with the high current carrying characteristic of tetrodes. The idling anode current also is only about one third of the usual amount; or, in other words, for a given valve the maximum power obtainable is greatly increased.

Power Supplies.—It will be noted that the schematic diagrams in many cases envisage more than the usual number of power supplies, especially when those for screens and grids are worked out, and particularly if d.c. connections are desired. This complication is not as great as may be imagined when voltage doubling circuits, the avoidance of mains transformers, and direct series running of the heaters are considered. Moreover, due to the large amounts of negative feedback used, the hum in the output is generally reduced so much that chokes need not be used in the supplies. In fact, such chokes are often undesirable, as the impedance of the h.t. supply to the output valves must be low compared with the load to avoid loss of power.

Practical Choice of Output Valves.-To return to

some earlier remarks, one of the biggest practical difficulties is simply that of obtaining output valves that can pass the necessary current. Here from the cost point of view more than one valve will almost certainly be required for each side of the output stage. Bearing in mind that certain valves can be obtained very cheaply, it may be more economical to use a large number of one valve rather than fewer of another. Valves made primarily for other purposes, such as television line scan or current stabilization, can also sometimes be of use. The resultant optimum load should, of course, be made as small as possible, though increased distortion due to mismatch may be removable by sufficient feedback.

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- ² V. Brociner and G. Shirley, "The Output-Transformer-less Amplifier," *Audio Eng.*, Vol. 36, June 1952, p. 21.
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B.B.C. Facts and Figures

OVER 20% of the 14,519 members of the B.B.C. staff are classified as being in technical engineering.

The total staff engaged exclusively on work for the television service is 3,700—about 25% of the whole.

Nearly 24% of the £11M spent during the last financial year on the sound services (excluding overseas transmissions) was devoted to engineering. Of this sum £237,269 went to the Post Office for rental of lines.

Of the £7M expended on television, 42% (£2.95M) was debited to engineering. Over half a million pounds was paid to the Post Office for lines.

The Post Office received, in all, nearly £1M from the B.B.C. for the rental of lines during the year ended last March.

During 1955-56 the gross licence income was £25.736M. Of this sum the Treasury took £2.75M and the Post Office retained £1.784M, leaving the B.B.C. £21.2M.

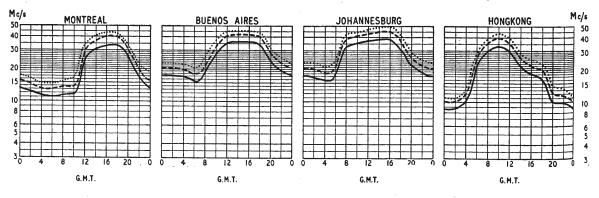
Thirty-nine high-power, short-wave transmitters and 177 aerials are used for the B.B.C.'s external services.

Over 20,000 hours of B.B.C. recordings were transmitted in 1955, and during the last five months of the year tape accounted for 70% of the total.

These facts and figures are culled from the "B.B.C. Handbook, 1957" (B.B.C. 5s), and the "Annual Report and Accounts of the B.B.C., 1955-56" (H.M.S.O. 6s).

SHORT-WAVE CONDITIONS

Prediction for February



THE full curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during February.

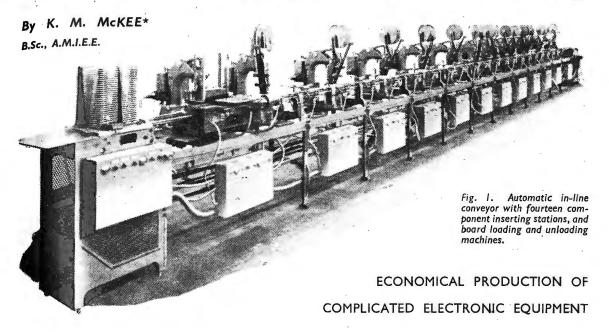
Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.

•••••• FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE FOR 25% OF THE TOTAL TIME

----- PREDICTED AVERAGE MAXIMUM USABLE FREQUENCY

FREQUENCY BELOW WHICH COMMUNICATION SHOULD
BE POSSIBLE ON ALL UNDISTURBED DAYS

Automatic Component Assembly



ITH the remarkable rate of growth of the electronic industry during the past twenty years it is not really surprising that the production side has had little time to investigate fundamentally new methods of circuit construction. Development engineers have been more than fully occupied incorporating in their designs improved components and rapidly developing circuit techniques, and at the same time in maintaining prices at a level which compared favourably with competitors.

The conventional form of the electronic circuit appears to have been derived from the larger and more robust electrical power installation, in so far as both consist essentially of independently manufactured units interconnected by wires. While this form of electronic circuit offers flexibility in construction, its reproduction in quantity involves a multitude of point-to-point wires, terminals, insulators and metal fittings. The result has been a high labour content and the existence of numerous possible sources of error in assembly. It is significant that on a normal assembly line in the radio and television industry one inspector is required for every five operatives.

A further difficulty lies in the fact that random effects due to minor differences in wiring layout make it hard to obtain consistent standards of performance, particularly where high frequencies are involved. Moreover, the variable and complex nature of the layout itself allows very limited scope for employing mechanical aid to improve the rate

and general efficiency of assembly.

that if industry is to meet the demand for an increased volume of more complicated electronic circuits at reduced cost, the basic form of circuits

Advent of the Printed Circuit.—It has become clear * Geo. Tucker Eyelet Co. Ltd.

would have to undergo a radical change in order to introduce the machine into every stage of production. Various schemes have been evolved using a variety of methods, several of which have involved producing complete electronic circuit assemblies from basic material. In this connection it has proved not at all easy to reproduce in the circuit elements that go to make up these assemblies the reliability and tolerances associated with orthodox components. After all, the characteristics of standard components have been built up as a result of years of development and production experience.

The form of assembly that has now gained wide acceptance is the combination of standard components and printed-wiring boards. Printed-wiring boards etched from copper-clad phenolic laminate have the great merit from the production standpoint that all conductor terminations are fixed in position and in a single plane. Above all, mechanized processing can be applied in turning out identical wiring boards, so forming a suitable basis for automatic assembly and subsequent mechanized operations.

It has been estimated in the report1 of a comprehensive survey conducted on behalf of the U.S. Navy that with assemblies produced in this way, the labour content is reduced thirty times in com-parison with conventional methods; in both cases production capacity and rate were assumed to be constant. To assist in achieving this remarkable saving in human effort, the report recommends the of automatic multi-station machinery for the assembly of standard components on the printed-wiring boards.

Automatic Assembly Machinery.—Fully automatic assembly machinery may be divided into two types; programmed single-station machines and multi-station in-line conveyor systems. Programmed single-station machines have a relatively low rate of output, are costly, and complicated in concept and action. Accurate programmed tapes are necessary; though once the tapes have been made, no other machine setting-up routine is necessary. The field of application lies chiefly in the assembly of batches of very short production runs.

The alternative is the multi-station in-line conveyor system where the intricate assembly operation is broken down into a number of easy stages, which in turn implies a series of simple and reliable mechanisms. Separate pre-set machines mounted on a transfer conveyor, each machine (or station) inserting a single type of component into the printed-wiring boards as they pass down the line. The machines are arranged to operate simultaneously so that on completion of each machine inserting cycle an assembled wiring board emerges at the end of the line. Although the movements of an in-line transfer conveyor could hardly be described as fast, surprisingly high output rates of

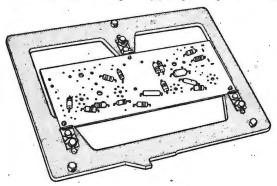


Fig. 2. Conveyor pallet complete with printed-wiring board and assembled axial-lead components.

Fig. 3. Simplified diagram showing the circulation of pallets on the forward and return conveyor belts. The electrical control system is illustrated comprising microswitch units and machine-control panels at each station, together with the conveyor-control panels on the end sections.

up to 1,200 assembled boards an hour can be regularly maintained. What is more, the conveyor may be extended to accommodate any number of additional machines without change of output rate; though a line of forty stations is now regarded as the practical upper limit.

Application in Industry.—Early in 1956, eighteen in-line transfer conveyors were in regular operation in the factories of leading United States electronic manufacturers. The most widely used system was that incorporating "Dynasert" automatic equipment. This machinery2 is now being developed and manufactured in this country, and is being made available to the British electronic industry by the Geo. Tucker Eyelet Company under the

"Dynasert" trade mark.

The first of these conveyors to be installed in the United Kingdom is at present in production on a commercial basis at the Ekco factory at Southend, and a second is shortly to be in operation in the works of another well-known British radio and television manufacturer. During the past year, as a preliminary step to full automatic assembly, six other manufacturers have purchased machines modified as manually operated bench machines. machines (which will later be described in greater detail) can readily be converted for use on a fully automatic conveyor.

Conveyor Operation.—The conveyor, Fig. 1, consists essentially of a transfer machine for transporting the printed-wiring boards, and bringing each in turn into an accurately located and firmly locked position beneath the inserting machine heads at every station. The phenolic laminate material of the printed-wiring board is unfortunately inclined to be dimensionally unstable, particularly after having been subjected to unavoidable temperature cycles during the etching and punching process. It has been found that mounting the board on a light alloy frame or pallet for assembly, as shown in Fig. 2,

COMPONENT

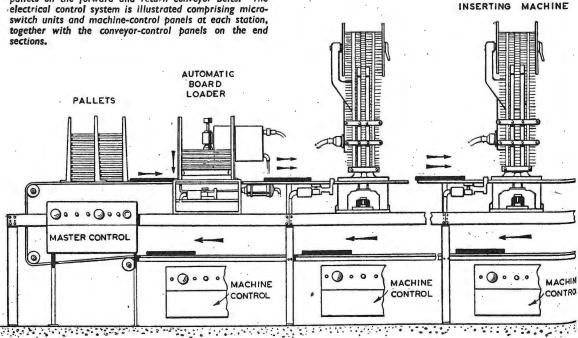


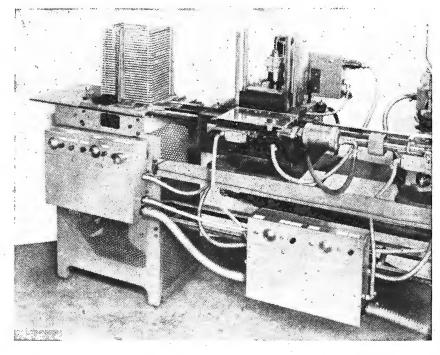
Fig. 4. Loading end of the conveyor with stack of empty pallets and loading unit for mounting printed-wiring boards on pallets prior to component assembly.

assists accuracy of location and enables a range of board shapes and sizes accommodated be without adjustment to the conveyor.

A simplified drawing illustrating the action of the conveyor and the circulation of the pallets is shown in Fig. 3. Pallets are stacked at the lefthand end of the conveyor and are released one at time on to a set of twin conveyor belts (see C in Fig. 5) which are continually in motion while the conveyor is in operation. At the next printed-wiring station,

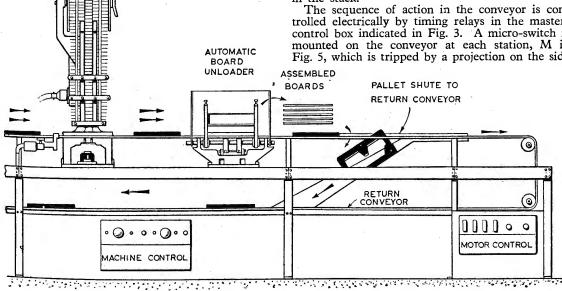
boards are automatically fed beneath the ram plate of a pneumatic press, Fig. 4, which thrusts the board downwards between the spring clips on the pallet. The board is firmly held on the pallet by steel pins which fit into accurately punched location holes at the edges of the board, Fig. 2. One of the holes must be round, the remainder may be drawn to allow for any change that may have taken place in the lateral dimensions of the board.

The pallet with the board is now released and transported by the conveyor belts to the first inserting station, where it is lifted from the belts and solidly locked beneath the machine head. On com-



pletion of the inserting cycle, the pallet is once more released. In this way, a series of pallets carrying printed-wiring boards proceeds down the line until the board-unloading station is reached at the distant end. Here the reverse of the loading process takes place, with the boards being pushed from the pallets by a ram press acting from underneath. Fingers on arms incorporated in the unloader unit lift the boards, which then slide off down a chute (not The empty pallets are tripped off the forward moving belts and descend by another chute to the lower return conveyor, where they pass back to the head of the line. It will be observed that only a small number of pallets is required to maintain circulation, plus a few extra to form a reserve in the stack.

The sequence of action in the conveyor is controlled electrically by timing relays in the mastercontrol box indicated in Fig. 3. A micro-switch is mounted on the conveyor at each station, M in Fig. 5, which is tripped by a projection on the side



of each pallet, Fig. 2, when the pallet reaches a The micro-switches are connected in a series circuit, so that only when every pallet has arrived at its station will the circuit be complete, and the timing relay actuated in order to allow the The circuit thus insertion cycle to commence. serves as an interlock to arrest action should a serious conveyor stoppage occur, in which event a signal light on the machine control box in question is illuminated to attract attention. Switches fitted to each machine control box enable units to be operated independently or isolated from the main control circuit.

Component Inserting Units.—Over 70% of the components used by the electronic industry for radio and television equipment are of the axial-lead type, and the vast majority of these are of the 4-W resistor size or smaller. In considering the design of automatic inserting machines, it was obviously essential to develop a rugged and reliable unit at reasonable price to handle this particular range. The No 1 "Dynasert" component inserting machine illustrated in Figs. 5 and 6 has proved to be capable of maintaining an insertion reliability of better than 99.8%.

A high insertion reliability of this order is vital if the operation of a line of automatic machines is to be a practical proposition. For instance, if the machine insertion reliability was only 95%, according to the laws of probability with ten machines in line, 40% of the assembled printed-wiring boards would contain a misplaced component. similar conveyor having a machine insertion reliability of 99.8%, the corresponding figure would be less than 2%. Good mechanical design and precision workmanship in making machine inserting head parts are required for high insertion efficiency.

Apart from axial-lead components, machines are in course of development for handling flat and disc radial-lead capacitors, and a wide variety of printed circuit valve bases. Another machine of this type is also being developed for inserting printed circuit sub-assemblies, consisting of a number of components mounted on a small printedwiring board. The sub-assembly may be made up on a conveyor with the manufacturer's own choice of components, or, alternatively, purchased as a packaged item from a component maker. Subassemblies when mounted vertically on a larger printed-wiring board introduce a third dimension to the layout, and their application in certain circuits leads to economy in board area.

A further range of conveyor machines has been planned for placing eyelets, tags and special terminals, which are often needed for interconnecting wiring between one printed-wiring board and another, and from them to the larger units used

with electronic equipment.

Axial-lead Component Machines.—Machines for axial-lead components will be made in three sizes, which depend mainly on the dimensions of the component body to be handled and the corresponding lead-hole spacing on the printed-wiring board. The two sizes of unit at present available will handle components of the smallest body length and a minimum body diameter of 32 in up to approximately lin body length and sin diameter. The third unit is being designed to insert components with maximum body dimensions of 2½ in in length and lin in diameter.

Also, a modified version of the axial-lead component machine has been developed for inserting jumper wires. Jumper wires are cut from a reel of plain or coated wire; then formed, inserted, and clinched in a similar manner to axial-lead components. Circuit designers have found jumper wires of special value in solving printed-wiring layout problems, as they are able to form a useful bridge connection between any two printed conductors.

For feeding purposes, axial-lead components are belted on reels before being loaded on to the machine. It is hoped that component manufacturers will soon make their standard axial-lead components available in belted form. "Dynasert" belting machines which straighten out slight bends and kinks in the leads and belt-up the components on reels, are soon to be available for use by both component and electronic equipment manufacturers. It is now generally accepted that reels of belted components with straight leads provide a most economical and efficient method of packaging for

all applications.

The machine shown in Figs. 5 and 6 is a typical example of the units used for inserting belt-fed components. Components on leaving the reel slide down the guide bars G, which are of sufficient length to allow reels to be changed without inter-fering with continuity of operation. The leads are then engaged either side of the body by grooves cut round the peripheries of two feed wheels, Fig. 7, and drawn under the inserting tools in the head. On air being released in to a pneumatic cylinder, Fig. 6, which is mounted in the frame, F in Fig. 5, the tools are thrust downwards towards the printedwiring board. In the process, excess length is trimmed from the leads, which are then formed through a right-angle at each end, and driven down and through the pre-punched holes in the board. The remaining wire trimmings are carried clear of the inserting mechanism by the tray T, Fig. 5, still attached to the tapes.

The pneumatic cylinder through a separate lever system raises two anvils, Fig. 7. The leads on emerging from the underside of the board are clinched over by the anvils in any pre-set direction, the ends being normally arranged to lie along the printed conductors, as indicated in Fig. 8. Experience of many millions of insertions in the U.S.A. has shown that leads clinched in this way greatly assist the formation of sound dip-soldered connec-

tions to the printed wiring.

Referring to Fig. 7, a resistor is represented in three of the stages of insertion and clinching.

The overall cut-off length

$$1'=L+D+d+2(t+s+c)$$

Substituting the values of dimensions in the brackets, which are recommended in the majority of cases:

$$1' = L + D + d + 2 \left(\frac{1}{16} + \frac{1}{8} + \frac{1}{16} \right)$$
 inches i.e., $1' = L + D + d + \frac{1}{2}$ inches.

Production Change-over.—Lack of flexibility in production change-over has in the past rendered some previous systems of automatic assembly impractical under modern working conditions. This could scarcely be said of the multi-station in-line system, where adjustment from one layout of printedwiring board to another usually takes less than six minutes per station.

Inserting machines are secured in position on

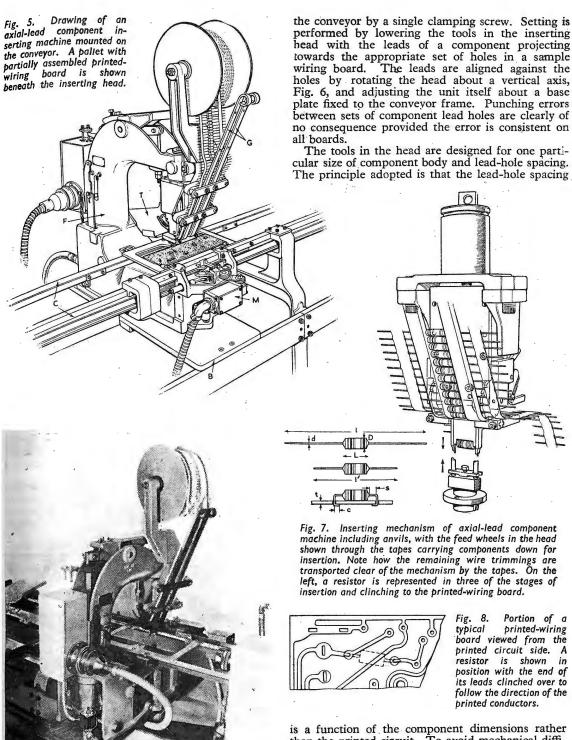


Fig. 6. Rear view of an axial-lead component inserting machine about to place a component. The inserting head has been rotated to align the component leads to the pre-punched holes in the printed-wiring board.

printed-wiring board viewed from the printed circuit side. A resistor is shown position with the end of its leads clinched over to follow the direction of the brinted conductors.

is a function of the component dimensions rather than the printed circuit. To avoid mechanical difficulties and the high cost associated with small variable setting devices, it was decided to produce tools for a range of standard hole spacings, with dimensions of the most used components in mind. Special tools can, however, be made for any leadhole spacings within the limitations imposed by the dimensions of the component and machine head.

A conveyor station may be converted from feeding

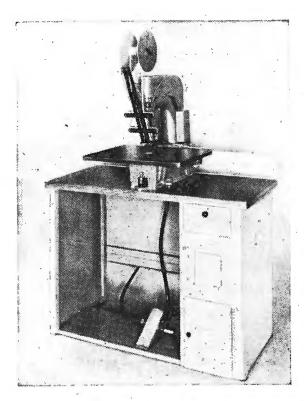


Fig. 9. Machine modified for bench operation showing the operator's foot control and with the safety cover guards in position on the machine.

one component to another of different dimensions by substitution of (i) the machine, or (ii) the inserting head, Fig 7, or (iii) the tools in the inserting head. The change-over in (i) and (ii) may be carried out by an unskilled operator in a matter of minutes, (iii) involves the services of a skilled mechanic for about 20 minutes.

Semi-automatic Production.—For short production runs conveyor machines may be mounted on a work table, Fig. 9, and employed as individual bench units. Only minor modification is required to the air control system of the conveyor model, for foot operation. Several such units could be adapted to cover a large percentage of the components needed for the circuits on most printed-wiring boards. Furthermore, by stripping the component-feed fittings from the head, a bench machine may be converted to a component cutting and forming unit, which is occasionally required by manufacturers for the preparation of special components prior to manual insertion.

Fig. 10 shows a close-up view of an optical location attachment fitted to the bench machine in Fig. 9. The arrangement projects two narrow beams of light on to a mirror, which reflects the light and produces two bright spots on the printed-wiring board. The board is quickly manipulated until the spots of light disappear down the required pair of holes, the foot pedal is depressed and the machine inserts and clinches the component. For a large batch of boards, it is preferable to use the arrangement as a setting-up aid for magnetic stops attached to the metal table, subsequently using the stops as a location jig for positioning the boards.

In comparison with manual performance of the same work, the machine shows a marked saving in time, and practically no operator fatigue has been experienced over long periods of operation.

Designing for Automatic Assembly.—Perhaps the most satisfactory feature of the new automatic production techniques from the point of view of the designer is that he has now almost complete control over the final performance of the circuit that he designs. If he succeeds in obtaining good results from a prototype in his laboratory, there is no reason why that performance should not be repeated in practically every model produced on the factory floor. On the other hand his responsibility is now much greater than it used to be. His design must be entirely free from error because it will often be an extremely costly matter to rectify a mistake once production has commenced. In the author's experience, most designers in the industry have welcomed the change and are only too willing to accept the increased responsibility entailed.

Many firms are now in the difficult stage of transition, where they are attempting to adapt designs and layouts that were intended for conventional wiring. This is nearly always a most difficult task to accomplish satisfactorily, and it is usually quicker to scrap the original layout and start again. In the United States, where now nearly 90 per cent of the mass-produced electronic equipment is on printed wiring, it has been found that three to five boards are most suitable for a television receiver, and one or two for a radio receiver. Hole-positional errors outside the tolerances required for automatic assembly, servicing problems, and breakages in punching, have turned the scales against the use of large-size wiring boards. It is now generally recognized in all fields of production engineering that the product must be designed with automatic production in mind.

Component manufacturers have also realized that parts developed for flexible wiring create serious mechanical problems in their assembly on printed circuits. The new tendency is for leads to become stouter and shorter, with larger and awkwardly shaped components fitted with snap-in type connections. Physical dimensions will have closer tolerances and a high degree of standardization among

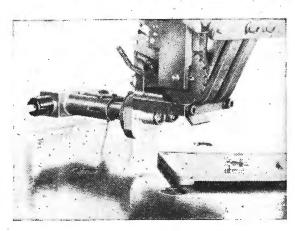


Fig. 10. Optical ocation attachment fitted to a bench machine to facilitate positioning of the printed-wiring board before insertion of a component.

various manufacturers' products will become increas-

ingly important.

Advantages for the Future.—It might well be asked what is on the credit side for the effort and capital that will be expended in changing from the old to the new manufacturing techniques. There is little doubt that substantial financial savings will be possible in production costs, not so much by considering any single process on its own, such as automatic assembly, but by evolving a complete flow-through production system commencing with mechanized processing of printed wiring and followed by automatic assembly, mechanized dipsoldering, and automatic electrical testing. Among other outstanding advantages are increased uni-

formity and reliability of the final product with far less inspection, rapid production change-overs and a massive production potential available to meet any emergency or unforeseen increase in demand.

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B.B.C. F.M. Transmitter Performance

SOME insight into the standards of quality specified by the B.B.C. and provided by the manufacturers of the transmitters now being installed for the v.h.f. service is given in papers* recently read at the Institution of Electrical Engineers and to be published in the Proceedings (1957, Vol. 104, Part B).

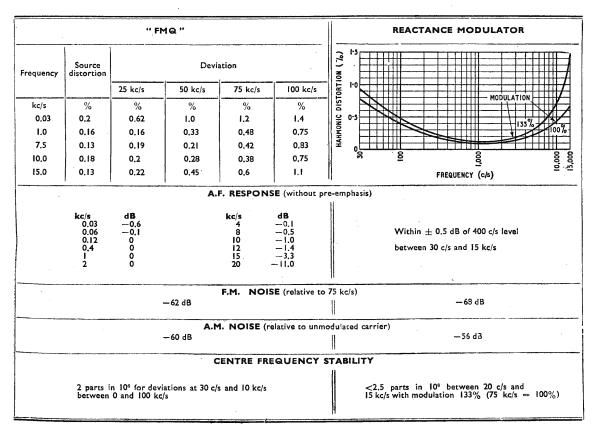
Two systems of modulation are being used: the transmitters supplied by Standard Telephones and Cables make use of balanced reactance-valve modulation of a free-running oscillator whose centre frequency

* "Frequency-Modulated Quartz Oscillators for Broadcasting Equipment." W. S. Mortley, A.M.I.E.E. "Frequency-Modulated V.H.F. Transmitter Technique," A. C. Beck, B.Sc.(Eng.), A.M.I.E.E., F. T. Norbury, A.M.I.E.E., and J. L. Storr-Best, B.Sc.(Eng.), A.M.I.E.E.

is controlled by reference, after frequency division, to a low-frequency crystal standard, and those supplied by Marconi's Wireless Telegraph Company rely on the "FMQ" system in which a low-frequency crystal-controlled oscillator is directly modulated by a balanced susceptance circuit† and followed by frequency multiplication.

The published figures for the performance of these two systems summarized below are expressed in different ways and are not directly comparable, but they indicate the high standards achieved in the frequency-modulated v.h.f. service.

† "FMQ." W. S. Mortley, A.M.I.E.E., Wireless World, October, 1951.



Limiters and Discriminators

2—Foster-Seeley Discriminator; Designing for Minimum Distortion

By G. G. JOHNSTONE*, B.Sc.

HE basic type of Foster-Seeley discriminator is shown in Fig. 1. A number of variants exist in practice, and these are discussed later. The audio output is the difference between the output voltages developed by the two diodes D1 and D2. At the centre frequency of the circuit, this output is zero, and the output swings above or below zero as the frequency of the applied signal shifts from its centre We shall assume initially that the diodes have 100 per cent rectification efficiency; the audio output is then equal to the difference of the peak values of the two r.f. voltages applied to the diodes.

Now let us analyse the r.f. side of the circuit. The transformation we shall employ is that shown in Fig. 2, the development of which was given in the first part of this article. The phase-difference transformer of Fig. 1 is then identical in performance with the arrangement of three tuned circuits shown. The "ideal" transformer T serves only to ensure that the current fed to its centre tap divides equally between the branches connected to its ends. We shall concentrate initially on the two tuned circuits connected between terminals 2, 3 and 2, 4. The parameters used are $x = 2Q_s df/f_o$ and $R_s/2$, the dynamic resistance of the tuned circuits. expression for x, df is the shift of the signal from its centre frequency f_o , whilst the value of Q_s is given by R_s/L_s . The centre frequency f_o is given by $f_o = 1/2\pi\sqrt{L_sC_s} = 1/2\pi\sqrt{L_pC_p}$. The special values of $x = \pm x_1$ give the displacement of the resonant frequencies of each of the two circuits from the centre frequency. In the circuit shown in Fig. 2, the resonant frequency of one tuned circuit is given by

^{*} B.B.C. Engineering Training Department.

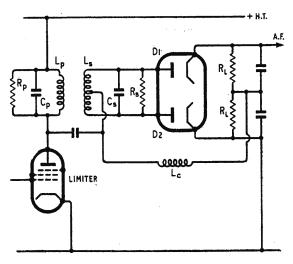


Fig. 1. Basic Foster-Seeley discriminator circuit.

$$\left(1+rac{M}{2L_p}
ight)^{-rac{1}{2}}f_o$$
 and that of the other by $\left(1-rac{M}{2L_p}
ight)^{-rac{1}{2}}f_o$. These values are equal approximately to $\left(1-rac{M}{4L_p}
ight)$

 f_o and $\left(1 + \frac{M}{4L_p}\right) f_o$ respectively, and hence

 $x_1 = 2Q_sM/4L_p = Q_sM/2L_p$. This can be rearranged into a more convenient form as follows.

$$x_1 = \frac{Q_s M}{2L_p}$$

$$= \frac{Q_s k \sqrt{L_s L_p}}{2L_p} \text{ where } k \text{ is the coupling coefficient}$$

$$= \frac{kQ_s}{2} \sqrt{\frac{L_s}{L_p}}$$

It is shown in the Appendix that if E_p and E_s are peak values voltages across the primary and secondary winding of the phase-difference transformer at resonance, then

 $x_1 = (E_s/2)/E_p$ With a constant-current input I/2 to each of the two tuned circuits connected between terminals 2, 3 and 3, 4 the difference between the peak r.f. voltages between terminals 2, 3 and 2, 4 is given by

$$E = IR_s/4 (a_1x + a_3x^3 + a_5x^5 \dots)$$

where

$$a_1 = 2x_1(1 + x_1^2)^{-3/2}$$

$$a_3 = x_1(2x_1^2 - 3)(1 + x_1^2)^{-7/2}$$

$$a_5 = \frac{1}{4}x_1(8x_1^4 - 40x_1^2 + 15)(1 + x_1^2)^{-11/2}$$

E is equal to the a.f. output provided that rectification efficiency is 100 per cent.

Except for the special case when the elements of the tuned circuit connected between terminals 1 and 2 become of infinite impedance, the input current, I, to the centre tap of the "ideal" transformer T is not constant. Its value can however be calculated.

If the input current to terminals 1, 2 is I_{in}, then I is equal to I_{in} less the current flowing in the tuned circuit connected between terminals 1 and 2. In order to simplify the treatment, we shall assume

that the Q-values of the two circuits of the phase difference discriminator transformer are equal. Additionally, we shall employ the relationship $p = L_s/L_p$. The equivalent circuit can then be

edrawn as shown in Fig. 3. The relationship between I and I_{in} is then given by

$$I = I_{in} rac{1 + {x_1}^2}{rac{j}{4} + {x_1}^2} \left[rac{1 + 2{x^2}rac{{(1 - {x_1}^2)}}{{(1 + {x_1}^2)^2}} + {x^4}rac{1}{{(1 + {x_1}^2)^2}}}{1 + 2{x^2}rac{{(1 - {n^2})}}{{(1 + {n^2})^2}} + {x^4}} rac{1}{{(1 + {n^2})^2}}
ight]^{rac{1}{2}}$$

where n = kOAt first sight it would appear that the output

WIRELESS WORLD, FEBRUARY 1957

for F.M. Receivers

In the first article of this series an equivalent circuit for the phase-difference transformer employed with the Foster-Seeley discriminator and the ratio detector was derived. This equivalent circuit enables these two forms of detector to be treated in the same manner as the Round-Travis circuit, already discussed, and in this part we shall discuss the Foster-Seeley circuit in greater detail.

voltage depends upon three variables, x_1 p and n. This is however, not true because

$$x_1 = \frac{1}{2} kQ \sqrt{\frac{L_s}{L_p}} = \frac{1}{2} n \sqrt{p}.$$

Thus there are only two independent variables. From the equation for I, it will be apparent that I is independent of x only if $x_1 = n$. This is the special case referred to above, when the inductance L' becomes infinite, capacitor C' becomes zero, and R' becomes infinite. In these conditions $\sqrt{p} = 2$ and hence $I = I_{in}$ as would be expected.

It was shown in Part I that the coefficient a_3 in the expression of the output voltage is zero when $x_1 = \sqrt{1.5}$. In the phase-difference transformer, the same conditions apply when $n = \sqrt{1.5}$, and $L_s = 4I$...

 $^{4}L_{p}$. In the general case, when x_{1} does not equal n, we can express I as a power series in x, as follows

$$I = I_{in} \frac{1 + x_1^2}{\frac{p}{4} + x_1^2} (b_o + b_2 x^2 + b_4 x^4 ...)$$

where

$$egin{align*} b_o &= 1 \ b_2 &= rac{1 - x_1^2}{(1 + x_1^2)^2} - rac{1 - n^2}{(1 + n^2)^2} \ b_4 &= rac{2x_1^2}{(1 + x_1^2)^4} + rac{1 - 4n^2 + n^4}{(1 + n^2)^4} - rac{1 - x_1^2}{(1 + x_1^2)^2} \cdot rac{1 - n^2}{(1 + n^2)^2} \ . \end{align}$$

Inserting this value for I in the expression for Egives

$$E = \frac{I_{in}R_s}{4} \cdot \frac{1 + x_1^2}{\frac{p}{4} + x_1^2} (c_1x + c_3x^3 + c_5x^5 \dots)$$

where

$$c_1 = (a_1 b_0)$$

 $c_3 = (a_1 b_2 + a_3 b_0)$
 $c_5 = (a_1 b_4 + a_3 b_2 + a_5 b_0)$

The distortion introduced is represented by the terms in x^3 and x^5 . To minimize distortion, therefore, it is desirable that the coefficients of these terms should be as small as possible. The dominant term is the coefficient of x^3 , and this can be made equal to zero by appropriate choice of parameters. For this condition, a_1 $b_2 + a_3$ $b_0 = 0$, and substituting values this gives

$$-x_1^2 (2x_1^2 - 3) (1 + x_1^2)^{-7/2} = 2x_1 (1 + x_1^2)^{-3/2} \times [(1 - x_1^2) (1 + x_1^2)^{-2} - (1 - n^2) (1 + n^2)^{-2}]$$

This reduces to

$$\frac{1}{(1+x_1^2)^2} = \frac{2(n^2-1)}{(1+n^2)^2}$$

The graph of x_1 plotted against n is given in Fig. 4. This shows that if n is less than 1, c_3 cannot equal zero. It also shows that if n is between 1 and 1.2 approximately, there is little margin for error in

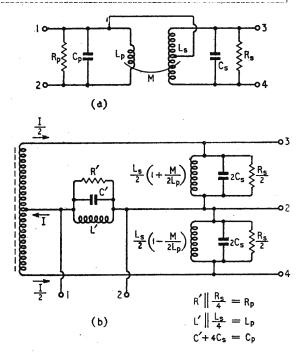


Fig. 2. (a) Phase-difference transformer, and (b) its equivalent circuit.

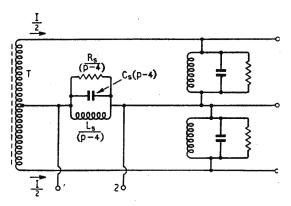


Fig. 3. Equivalent circuit of phase-difference transformer for $Q_s = Q_p$; $p = L_s/L$.

adjustment of n, since x_1 is varying rapidly. The minimum value of x_1 occurs when $n = \sqrt{3}$; at this value $x_1 = 1$. Above $n = \sqrt{3}$, the slope of the curve is positive. This is of some importance, since x_1 is itself proportional to n. If n departs from its correct value, it is desirable that x_1 should change in the same sense to minimize the value of c_3 .

The graph of x_1 plotted against n does not indicate any specific optimum value for n and x_1 . However,

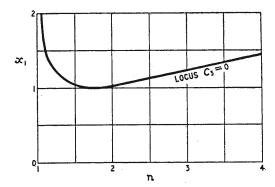


Fig. 4. Graph of x_1 against n for $c_3 = 0$.

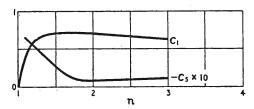


Fig. 5. Variation of c_1 and c_5 with n, for $c_3 = 0$.

this can be found by considering the coefficient of x^5 in the expansion. The value of this coefficient is plotted in the Fig. 5; at each value of n the value of x_1 is that which makes $c_3 = 0$. Additionally the graph shows the value of the coefficient of x; minimum distortion occurs when the ratio of the coefficient of x^5 to that of x is minimum. The optimum values would appear to be near n = 2.0. Consider n = 2; in this region the coefficient of x^5 is approximately -0.008. From Fig. 4, $x_1 = 1$ approximately.

From $x_1 = \frac{1}{2} n\sqrt{p}$, p = 1, i.e. the two tuned circuits of the phase difference discriminator transformer are equal. From Fig. 5, the coefficient of x is 0.7 approximately.

The expression for the audio output voltage is then $E = 0.4 I_{in}R_s (0.7x -0.008x^5 \dots)$

The range of validity of this expression can be seen from Fig. 6. In this graph are two curves. Curve (a) is that obtained from graphical orthodox the solution for the Foster-Seeley response curve, which does not give any precise means of evaluating distortion; curve (b) is that derived from the expression above. It will be seen that the two curves are in very close agreement up to the value of x=1.4. This then is the limit of validity of the expansion above.

To evaluate the distortion present with the circuit constants chosen, consider an input signal $x_s \cos \omega_a t$. The a.f. output is then

 $E = 0.4 I_{in} R_s \{0.7x_s \cos \omega_a t - 0.008 (x_s \cos \omega_a t)^5\}$ We can expand $\cos^5 \omega_a t$ by means of the identity

$$\cos^5 \theta = \frac{1}{16} \left\{ \cos 5 \theta + 5 \cos 3 \theta + 10 \cos \theta \right\}$$

giving

$$E = 0.4 I_{in} R_s \left\{ (0.7x_s - 0.005x_s^5)\cos \omega_a t - 0.0025 x_s^5 \cos 3\omega_a t - 0.0005 x_s^5 \cos 5\omega_a t \right\}$$

The reduction of the fundamental frequency component is negligible for the range of values of x_s of interest, i.e. $x_s < 1$. The percentage of third harmonic

distortion is thus given by
$$\frac{0.0025 \times 100 \times x_s^4}{0.7}$$
. At

 $x_s = 1$, this is 0.35 per cent. By employing a smaller value of x_s a lower value of distortion is obtained, the distortion decreasing with x_s^4 .

Consider a broadcast signal, with a deviation of 75 kc/s. If it is desired to operate the discriminator with 75 kc/s corresponding to $x_s = 1$, the parameters of the circuit are determined by x = 2Q df f_o . With df = 75 kc/s at x = 1, and a centre frequency f_o of 10.7 Mc/s, the value of Q is 71. If we assume the two tuned circuits of the phase-difference transformer each employ a tuning capacitor of 50 pF, the dynamic resistance (R_s) is $22 \text{ k}\Omega$. The input current I_{in} , is the peak value of the fundamental frequency component in the output of the preceding limiter stage; a typical value is 1 mA. The peak audio output is given by $0.28 I_{in}R_s$, and in this example is 6.2 volts approximately.

It can be seen, from the expression for E, that the audio output is proportional to $I_{in}x$, where I_{in} is the input current, and x is a measure of the frequency

(Continued on page 73)

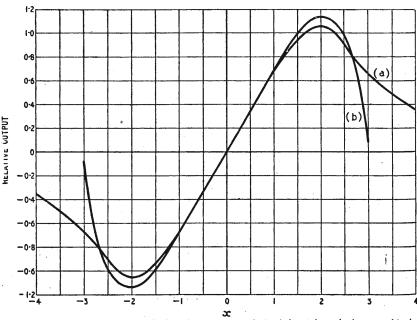


Fig. 6. Response of Foster-Seeley discriminator derived by (a) orthodox graphical solution (b) expansion given in text.

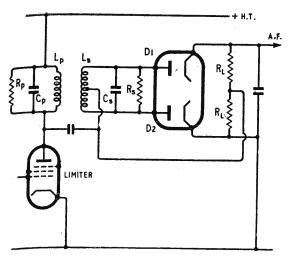


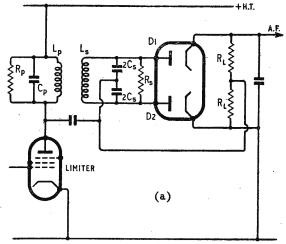
Fig. 7. Alternative form of Foster-Seeley discriminator.

shift. If there is amplitude modulation present, the magnitude of I_{in} varies, but if x=0, i.e. the signal is at the centre frequency there is no output due to a.m. For any other value of x, i.e. if the signal is mistuned or frequency-modulated, there is an output due to the amplitude modulation. Because the output is proportional to $I_{in}x$, the a.m. and f.m. signals are multiplied together and there is complete cross-modulation. Thus a Foster-Seeley discriminator must be preceded by a limiter stage.

To complete the survey of the Foster-Seeley circuit, there are a number of practical points to be considered. The first of these concerns the diode load resistors. These should not be too large, as otherwise "diagonal clipping" can occur. Briefly this happens if the input to one diode falls rapidly. If the time constant of the load circuit is too great, the cathode potential cannot fall sufficiently quickly, and the diode may be cut off. For this reason, the diode load resistors are usually limited to $100~\mathrm{k}\Omega$, and the shunt capacitors to $50~\mathrm{pF}$.

This relatively low value of diode load in turn means that the damping imposed in the tuned circuits cannot be neglected. The input resistance of each diode at r.f. is $R_L/2\eta$, where R_L is its load resistance and η is the rectification efficiency. In

the equivalent circuit of Fig. 3, the relationship between the resistances of the two tuned circuits connected between terminals 2, 3 and 2, 4 and that connected between terminals 1, 2 then differs from that postulated. The condition can, however, be reestablished if an additional resistor equal to $R_L/(p-4)$ is connected between terminals 1 and 2, i.e. across the primary winding of the original circuit. However, the values of p employed in practice are often less than 4, implying that a negative resistance is required. This obviously cannot be realised in practice. It suggests, however, that the Q values can be equalized if the secondary Q value without the diodes connected is lower than the primary Q value. Given equal initial Q values, a



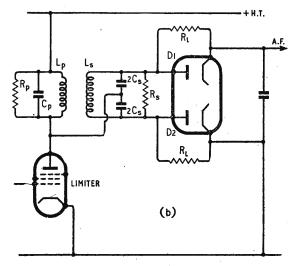
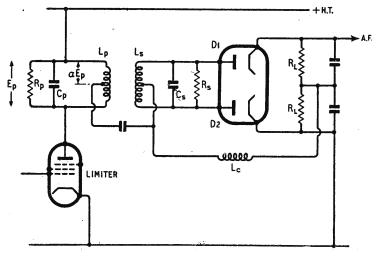


Fig. 8. Further alternative forms of Foster-Seeley circuit, with the secondary circuit centre taps produced by the capacitors.

Fig. 9. Foster-Seeley discriminator with tapped primary circuit.



resistor can be connected in parallel with the secondary circuit to achieve this result. Its value can be calculated if it is remembered that the damping imposed on the primary circuit is $R_L/4\eta$, whilst that imposed on the secondary circuit is R_L/η . In the example considered above, with equal primary and secondary circuit, the resistance would be $R_L/3\eta$.

It is common practice to omit the r.f. choke $L_{\rm o}$ shown in Fig. 1, giving the circuit of Fig. 7. In this case it must be remembered that there is then additional damping equal to $R_{\rm L}/2$ imposed on the primary circuit, since the primary circuit can "see"

the two resistors R_L in parallel.

An alternative form of Foster-Seeley circuit is that shown in Figs. 8(a) and 8(b). In the arrangement, the secondary centre-tap is provided by the capacitors. Because there is no longer d.c. continuity for the diodes of Fig. 8(b), the diodes are shunt-fed. The damping imposed on the secondary circuit is the R_L/η in parallel with $2R_L$, whilst that

on the primary is $R_L/4\eta$ in parallel with $R_L/2$. In one other variant the primary as well as the secondary is tapped, as shown in Fig. 9. If the ratio of voltage tapped off to the total primary voltage is a, then this circuit can be analysed by replacing the primary circuit by a single un-tapped circuit, with parameters as follows:

This circuit has one particular advantage. If $a = \frac{1}{2}$, i.e. the primary circuit is centre-tapped, the damping applied to primary and secondary circuits by the diode loads is automatically equalized. In such a circuit, the value of p in the equivalent circuit is equal to 4, i.e. the analysis simplifies to the case, when the tuned circuit in parallel with the input terminals vanishes. For this condition of operation, the optimum value of $n = \sqrt{1.5}$.

APPENDIX

The equivalent diagram for two circuits coupled by mutual inductance is shown below. The circuit equations are

where
$$egin{array}{ll} \mathbf{E}_s &= j\mathbf{X}_{cs}i_s \\ \mathbf{E}_p &= j\mathbf{X}_{cp}(i-i_p) \\ 0 &= \mathbf{Z}_pi_p - j\omega\mathbf{M}i_s - j\mathbf{X}_{cp}i \\ 0 &= \mathbf{Z}_si_s - j\omega\mathbf{M}i_p \end{array}$$
 where $egin{array}{ll} \mathbf{X}_{cs} &= 1/\omega\mathbf{C}_s \\ \mathbf{X}_{cp} &= 1/\omega\mathbf{C}_p \end{array}$ $\mathbf{Z}_p &= j\mathbf{L}_p\omega + rac{1}{j\omega\mathbf{C}_p} + r_p \end{array}$ $\mathbf{Z}_s &= j\mathbf{L}_s\omega + rac{1}{j\omega\mathbf{C}_s} + r_s \end{array}$

In the region near the resonants frequency (f_o) of the two circuits, $i_p > i$. Then

$${f E}_s = -j {f X}_{cp} {f X}_{cs} rac{\omega {f M}}{{f Z}_p {f Z}_s + \omega^2 {f M}^2} i \ {f E}_p = {f X}_{cp}^2 rac{{f Z}_s}{{f Z}_s {f Z}_s + \omega^2 {f M}^2} i \ {f E}_p$$

Near resonance $\omega L_s - 1/\omega C_s$ is approximately equal to $2L_s\delta\omega$, where $\delta\omega$ is the departure from $\omega_o(=2\pi f_o)$. Similarly, $\omega L_p - 1/\omega C_s = 2L_p\delta\omega$. This gives

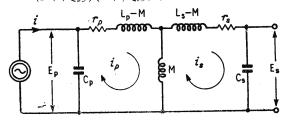
Similarly,
$$\omega L_p - 1/\omega C_s = 2L_p \delta \omega$$
. This gives
$$E_s = j \mathbf{X}_{cp} \mathbf{X}_{cs} \frac{\omega \mathbf{M}}{(r_p + 2j\mathbf{L}_p \delta \omega) (r_s + 2j\mathbf{L}_s \delta \omega) + \omega^2 \mathbf{M}^2} \cdot i$$

$$E_p = \mathbf{X}_{cp}^2 \frac{r_s + 2j\mathbf{L}_s \delta \omega}{(r_p + 2j\mathbf{L}_p \delta \omega) (r_s + 2j\mathbf{L}_s \delta \omega) + \omega^2 \mathbf{M}^2} \cdot i$$
Let
$$\frac{\mathbf{L}_p \omega_o}{r_p} = \frac{\mathbf{X}_{cp}}{r_v} = \mathbf{Q}_p; \quad \frac{\mathbf{L}_s \omega_o}{r_s} = \frac{\mathbf{X}_{cs}}{r_s} = \mathbf{Q}_s; \quad y = 2\frac{\delta \omega}{\omega_o} = 2\frac{\delta f}{f_o}$$

$$\frac{r_p}{r_p} = \frac{r_v}{r_v} = Q_p; \quad \frac{r_s}{r_s} = \frac{r_s}{r_s} = Q_s, \quad y - 2\omega_o$$
and $n = k \sqrt{Q_p Q_s}$ where $k = M/\sqrt{L_p L_s}$

$$E_s = \frac{-iQ_p X_{cp}}{(1 + iQ_p y)(1 + iQ_s y) + n^2} \cdot kQ_s \sqrt{L_s/L_p}. i$$

$$E_v = \frac{Q_p X_{cp}}{(1 + iQ_p y)(1 + iQ_s y) + n^2}. (1 + jQ_s y). i$$



whence
$$\mathbf{E}_s = rac{-j \, k \mathbf{Q}_s \, \sqrt{\mathbf{L}_s/\mathbf{L}_p}}{1 + j \mathbf{Q}_s \, y} \, \mathbf{E}_p$$

let $p = L_s/L_p$, and y = 0, i.e., the circuits are at the resonant frequency. Then if $|E_s|$ is the magnitude of E_s , and $|E_p|$ is the magnitude of E_p ,

$$\frac{1}{2} k Q_s \sqrt{p} = (\cdot |E_s|/2)/|E_p|$$

Finally if
$$Q_p = Q_s$$
, $Q_p X_{cp} = R_p$, and $Qy = \frac{2Q\delta\omega}{\omega_o} = x$

$$E_p = \frac{R_p(1+jx)}{(1+jx)^2 + n^2}i$$

In the equivalent circuit discussed in the text, this voltage is that applied to the tuned circuit connected between terminals 1 and 2 when $i = I_{in}$. In the text $p = L_s/L_p = R_s/R_p = C_p/C_s$. Thus the tuned circuit between terminals 1 and 2 consists of three branches in parallel; (a) inductor $L_s/(p-4)$; (b) capacitor $(p-4)C_s$; (c) resistor $R_s/(p-4)$ (see fig. 3). Then the current I' flowing into the three elements is given by

flowing into the three elements is given by
$$\begin{split} \mathbf{I}' &= \mathbf{E}_p \mathbf{Y} \\ \text{where } \mathbf{Y} &= (p-4)|j\omega \mathbf{L}_s + j\omega \mathbf{C}_s(p-4) + (p-4)/\mathbf{R}_s \\ &= (1|j\omega \mathbf{L}_s + j\omega \mathbf{C}_s + 1/\mathbf{R}_s) \, (p-4) \\ &= \frac{p-4}{\mathbf{R}_s} \, (1+jx) \end{split}$$
 Thus
$$\mathbf{I}' = \frac{\mathbf{R}_r (1+jx)}{(1+jx)^2 + n^2} \cdot \frac{p-4}{\mathbf{R}_s} (1+jx) \, \mathbf{I}_{in}.$$

$$= \frac{p-4}{p} \, \frac{(1+jx)^2}{(1+jx)^2 + n^2} \mathbf{I}_{in}$$

But the current I fed to the centre-tap of transformer T is equal to $I_{in}-I'$. Hence

Equal to
$$I_{in} = 1$$
. Hence
$$I = I_{in} \left\{ 1 - \frac{p-4}{p} \cdot \frac{(1+ix)^2}{(1+jx)^2 + n^2} \right\}$$

$$= I_{in} \frac{pn^2 + 4(1+jx)^2}{pn^2 + p(1+jx)^2}$$

$$= I_{in} \frac{4}{p} \frac{x_1^2 + (1+ix)^2}{n^2 + (1+jx)^2}$$

$$= I_{in} \frac{1 + x_1^2}{\frac{p}{4} + x_1^2} \cdot \frac{1 - x^2/(1+x_1^2) + 2jx/(1+n^2)}{1 - x^2/(1+n^2) + 2jx/(1+n^2)}$$

If |I| is the magnitude of I and |I_{in}| that of I_{in} then $|I| = |I_{in}| \frac{1 + x_1^2}{\frac{p}{4} + x_1^2} \begin{cases} \frac{1 + 2x^2(1 - x_1^2)/(1 + x_1^2)^2 + x_1^2}{1 + 2x^2(1 - n^2)/(1 + n^2)^2 + x_1^2} \end{cases}$

WIRELESS WORLD, FEBRUARY 1957

In the absence of a really cheap and simple colour display device for domestic receivers—on which the success of colour television so much depends—it may be considered somewhat profitless to discuss the question of transmission systems. There is, however, one point of view which should be heard. This argues that the system should not be tailored to fit the display device (as with the N.T.S.C. system and the three-gun shadow-mask c.r.t.) but should be made as perfect as possible in the expectation that display and receiving equipment will eventually be developed to match it. Such is the theme of this article-

N.T.S.C. Colour Information

Where the System Fails Through Expediency

By E. L. C. WHITE,* M.A., Ph.D., M.I.E.E.

ROM the vast body of work that has been done on colour television in the last few years a number of principles have emerged which now find very wide acceptance.

First is the idea of "compatibility." This means that the colour signals must be sufficiently similar in form to existing monochrome standards to give good black-and-white pictures, with no untoward effects, on monochrome receivers of normal design. Thus colour can be added to selected programmes, and the proportion increased as warranted by the increasing numbers of colour receivers.

Secondly, it has been recognized that a compatible system can be achieved by rearranging the primary red, green and blue signals into three other signals, one of which is representative of the brightness and therefore contains some of each of red, green and blue in suitable proportions. This brightness signal has the synchronizing pulses added to it, and is all that is needed for monochrome receivers.

The third principle is that the other two signals should only carry the colour information, as distinct from the brightness, and need only have a bandwidth which is a fraction of that of the brightness, or "luminance," signal.

Fourthly, there is the technique of adding the narrow-band colour signals, in the form of modulated sub-carriers, to the brightness signal within its normal frequency band. By adopting special frequency and phase relationships of the sub-carrier relative to the line scan, the objectionable effects of dot pattern on monochome receivers and loss of detail resolution on colour receivers and loss of detail resolution on colour receivers can be minimized. This technique has been, and still is, the subject of much controversy. In spite of its drawbacks, it will probably have to be accepted because of the scarcity of bandwidth in the radio-frequency spectrum available to television.

The fifth principle is the method of carrying the colour information, consisting essentially of two independent variables, on a single sub-carrier, by simultaneous phase and amplitude modulation. This is the subject which will be discussed here, with particular emphasis on the exact nature of the information carried.

As the N.T.S.C. system is now generally well

known, it is a useful standard of comparison. Its salient features have already been discussed in Wireless World† but a short recapitulation of some of the relevant points will be useful here.

At the transmitting end, the primary red, green and blue signals from the camera, after individual gamma correction, are applied to proportional adding circuits to form three other signals E_{γ} , E_{I} and E_{Q} . (Here the tick indicates that the signals are not linear but are tormed from gamma-corrected primaries. This gamma correction is to compensate for the non-linear light output characteristic of the receiver c.r.t., which usually follows a power law, so that the correction has to be an inverse power law). E_{γ} is the luminance signal already mentioned, while E_{I} and E_{Q} are known as "chrominance" signals.

 $E_{\rm I}'$ and $E_{\rm Q}'$ are used to modulate two orthogonal phases of a sub-carrier. Its frequency is an odd multiple of half the line scan frequency, to give dot interlace. The final output, which also includes sync pulses and a colour sync "burst," is formed by adding to $E_{\rm Y}'$ the modulated sub-carrier. The signals are band limited in varying degrees, on principles already discussed in the previous Wireless World articles.

The Colour Sub-Carrier

An important feature of the system is that the vector diagram of the colour signal is very similar in form to the standard chromaticity diagram (the international reference frame for colour specification), with the origin shifted to white. This is shown in Fig. 1, where the sub-carrier vector is superimposed on the chromaticity diagram. Hue becomes related to phase, and saturation to amplitude, and for white the sub-carrier has zero amplitude.

Another feature, and one which is not always realized, is that the amplitude of the colour signal is not dependent only on saturation, but also on

‡For an explanation of chromaticity and the chromaticity diagram se e "Colour Fundamentals" in the Aug. 1956, issue, p.363.

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brightness. This is a notable departure from the third principle enumerated above, and is, in fact, the main reason why the nature of the colour information transmitted by the N.T.S.C. system is open to

According to simple philosophy, the information should be pure chromaticity. This means dominant wavelength (or hue) and purity (or saturation), and corresponds to the familiar concept of colour quality, as distinct from quantity of light. chromaticity information can be given, for example, by the x and y co-ordinates in Fig. 1, or suitable linear transformations of them. These are essentially functions of ratios, as already explained in Wireless World,‡ and are independent of brightness.

The signals transmitted by the N.T.S.C. system, however, are not truly representative of chromaticity. They are "colour-difference" signals, as mentioned in previous articles† and have been given the name "chrominance" by the N.T.S.C. to distinguish

them from the idea of pure chromaticity.

There is another way in which the N.T.S.C. system departs from straightforward principles. arises from the process of gamma correction. already mentioned, the transmission system corrects for the non-linear electron-gun characteristics, or gamma, of the three-gun picture tube by interposing stages with an inverse power law between the linear signal sources and the proportional adding circuits.

From the short view, the reason is sound enough, as the alternative is to put the gamma correction immediately before the picture tube, which is uneconomical because it would be necessary in every receiver. However, the technique has several disadvantages, especially near saturated colours, and these are: (1) loss of luminance detail; (2) the system is no longer "constant-luminance," i.e. noise and interference in the sub-carrier band will produce brightness fluctuations as well as colour variations; (3) there is a non-linear relationship between the sub-carrier signal and the reproduced chromaticity,

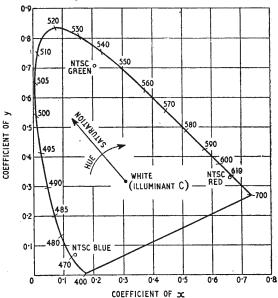


Fig. 1. Standard chromaticity diagram, with spectrum wavelengths in millimicrons and N.T.S.C. primaries. The vector representing the colour sub-carrier is superimposed to show its relationship.

which, for example, renders the hue near the complementary colours much more critical with regard to sub-carrier phase inaccuracy than is that near the primary colours.

Fig. 2 shows how the loss of luminance detail arises near saturated colours. Gamma is taken to be 2.0—that is, a square-law characteristic. effect of the narrow-band chrominance circuits on a step waveform is simplified by showing the output as a ramp function.

For a transition from red to blue at constant luminance, it will be seen that the square-law effect of the three-gun display-tube causes a dip in the displayed luminance. More striking, perhaps, is the blurring of the edge in a transition from dark blue to light blue. This is not due to the non-linearity, but to the fact that the colour signal is chrominance or colourdifference rather than chromaticity or colour ratio. Thus the major portion of the brightness change is carried over the narrow-band chrominance signal.

American Modifications

The solution to all these problems is to send a pure chromaticity signal for the colour information, but first some palliatives suggested by the N.T.S.C. will be mentioned.

The first arrangement entails a correction to the transmitted signal which removes the loss-of-detail fault shown at the bottom of Fig. 2. Unfortunately this leads to poor compatibility since the correction is not required by monochrome receivers, and on these the edges would be unnaturally emphasized.

The next step considered by the N.T.S.C. was to see what could be done at the cost of introducing one non-linear circuit in a three-gun receiver. This also modified only the luminance signal. It has the advantage of being correct for monochrome receivers, but still gives non-linear colour-difference signals on the sub-carrier.

Finally, schemes were considered in which more than one non-linear circuit might be required in an This led to the use of a true ideal receiver. luminance signal, as mentioned above, and a linear type of colour signal. Three possibilities were considered.

In the first, the chromaticity information was to be transmitted linearly in terms of the sub-carrier amplitude. Because of possible objection to a large sub-carrier amplitude at low brightness levels (giving, for example, reduction of contrast on monochrome receivers) a second alternative was discussed in which a linear chrominance signal was to be used. Besides being no longer a ratio type of system, this went to the other extreme, and the N.T.S.C. favoured a compromise in which chromaticity multiplied by a gamma-corrected luminance signal was used.

In general the receiving equipment required to take advantage of these modifications is either somewhat complex and expensive, or, if simplified, has a tendency to introduce new faults, such as

errors in brightness and hue.

From the above discussion, then, it emerges that the N.T.S.C. system, as at present practised, is tailored to give good large-area colour rendering on a three-gun type of tube driven from a receiver having no intentionally non-linear circuits, and that, nevertheless, the system has a number of failings which are probably more easily observed than colour inaccuracies would be.

While it is essential to aim at a low-cost receiverthe absence of which is at present the main brake on the development of colour television-it may be questioned whether in the long run the three-gun tube is the best basis for this. Not only do three guns add to the expense of manufacture, but they bring with them all the problems of registration, which are by no means solved in existing designs, and necessitate two extra wideband video output stages capable of providing about 100 volts swing and a stable black level.

Fundamentally what is required is a single electron gun controlled by a brightness signal, and a screen consisting of a mosaic of differently coloured phosphor dots or strips, with some mechanism which ensures that the spot only excites appropriately coloured dots or strips depending on the colour signal.

Two types of single-gun tube have been successfully demonstrated using this broad principle. One is the Chromatron, or "Lawrence" tube, using a beam deflecting grid near the screen (see July, 1953, issue, p. 329) and the other is the Philco beamindexing tube, in which the colour signal is applied to the gun in accordance with the position of the beam across the phosphor strips (see January, 1957, issue, p. 2). Both of these tubes have been used on the N.T.S.C. system. The receivers have needed somewhat greater complexity than those for the three-

gun tube, but this is largely due to the N.T.S.C. system being tailored to suit the last-mentioned.

Consider the basic requirements of these single-gun tubes, in which the beam excites the three primary phosphors sequentially and the colour is controlled by gating the beam on and off at appropriate times. For peak-white the brightness signal is a maximum, and it is reasonable to adjust the phosphor efficiencies so that the desired reference white (e.g., Illuminant C on Fig. 1) is obtained when each phosphor is equally excited. The symmetry of this arrangement necessitates a brightness signal which is

given by
$$E_{\rm B}=\frac{E_{\rm B}+E_{\rm G}+E_{\rm B}}{3}$$
. The symbol $E_{\rm E}$

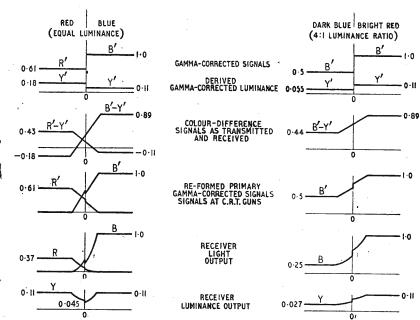
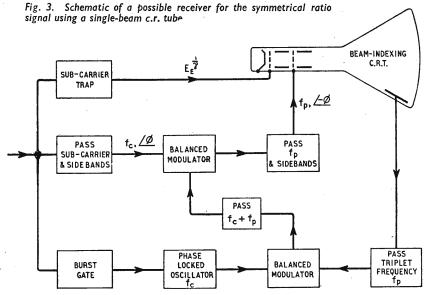


Fig. 2. Illustration of how gamma-correction causes loss of luminance detail near saturated colours. Read from top (transmitting end) to bottom (receiving end).



is used because this "brightness" corresponds more to total energy than to total luminosity. The signal $E_{\mathbf{g}}$ is preferably gamma-corrected at the transmitter for the receiver electron-gun characteristic, and is transmitted instead of the signal $E_{\mathbf{g}}$ ' in the N.T.S.C. system.

To accompany this "symmetrical" brightness signal, a symmetrical colour signal is also required at the receiver. Such a signal could consist of a subcarrier modulated in three different phases 120° apart by the three linear colour components $E_{\rm R}$, $E_{\rm G}$ and $E_{\rm B}$.

If the tube has a single beam control electrode, as in the existing Philco beam-indexing tube, then the colour signals, if they are to be used with the

minimum of processing, must be such that when added to the brightness signal in correct proportion they form the complete brightness and gating signal to be applied to the tube. This implies that the colour signal amplitude is non-linearly dependent on brightness, as in the N.T.S.C. system. Such a system, therefore, has the faults of the N.T.S.C. system, and in addition is not a constant-luminance system even near white.

Schemes are possible in which the colour signals gate or select the beam after it has been generated under the control of the brightness signal. Such schemes, in which the phosphor-selecting signals depend on hue and saturation, but not on brightness, require a signal which is both symmetrical, as explained immediately above, and a ratio signal as described earlier.

Symmetrical Ratio Receiver

In Fig. 3 is shown how simple, at least in principle, could be a receiver for a beam-indexing tube with this "symmetrical ratio signal." The regenerated subcarrier frequency f_c , locked in phase to the colour sync burst, is mixed with the beam index frequency f_p to produce $f_c + f_p$. This is again heterodyned with the colour signal E_c of frequency f_c and instantaneous phase ϕ , the lower side-band giving f_p , phase $-\phi$. Thus we now have a signal of beamindex frequency, of amplitude proportional to E_c and phase $-\phi$, which can be used to turn on the beam, to an amount controlled by the gamma-corrected E_B , at the appropriate times to excite the desired combination of phosphors. The gun would require two independent control electrodes (e.g. a hexode gun), giving a direct multiplication of the two signals.

A linear ratio system such as the N.T.S.C. modified system mentioned above preserves detail brightness resolution under all conditions, and the hue variation with phase of sub-carrier is independent of saturation. The latter feature is true also of the symmetrical ratio system, but as it is not a constant-luminance system, there is luminance distortion at any edges where the chromaticity changes. It is, however, less subject to brightness noise originating in the colour transmitting channel than was, for example, the original RCA dot sequential system, which was similar in many respects except that it was not a ratio system.

Thus the average amplitude of the colour signals is greater in the proposed system, and the extra brightness noise, especially in the medium brightness areas, where noise is most visible, is not expected to be noticeable. In comparing it with the present N.T.S.C. system, it must be remembered that the last-mentioned is only nominally a constant-luminance system, and departs radically from this principle in or near saturated regions.

The fact that the symmetrical ratio system is not constant-luminance in operation can be shown to be due entirely to the use of gamma-corrected $E_{\rm E}$ instead of gamma-corrected $E_{\rm Y}$ for the brightness signal, and does not depend on the form of the subcarrier modulation so long as this is linear. Another effect of using gamma-corrected $E_{\rm E}$ is that the luminance displayed on monochrome receivers will be incorrect, but as the signal is still panchromatic the resulting monochrome picture is likely to be perfectly acceptable. Incidentally, it is important to note that good results can be obtained on a three-gun tube with this system.

Summarizing the differences between the N.T.S.C. system and the symmetrical ratio system, the fact that with a three-gun tube decoding is performed outside the tube permits independent weighting of the red, green and blue signals, while for single-gun tubes with internal decoding symmetrical signals are preferable. In the N.T.S.C. system the facility of independent weighting is used to construct a nominally constant-luminance system, but because the signals thus weighted are not linearly related to the light intensities, constant luminance is only achieved near white.

It is clear from the discussion above that there is no perfect combination of system and tube. In these circumstances it might be best to forget about expediency, which is always a doubtful guide, and look for a system which is itself as perfect as possible, in the expectation that tube and receiver design will then develop along the necessary lines to match the system, so as to give overall optimum economy and performance.

It may be assumed that such a system would have a true luminance signal, corrected for normal gamma, but it is not so obvious what the colour signal ought to be, beyond the fact that it should be linearly related to chromaticity. It is preferably symmetrical in the sense that any of the three saturated primaries should give equal amplitude, and preferably white is zero, but these two conditions cannot be realized simultaneously unless the divisor used to form the ratios is $E_{\rm R}$ and not $E_{\rm Y}$.

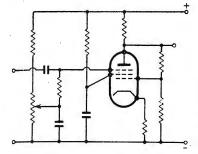
Thus no firm suggestion is made, but a useful purpose will have been served if it is now appreciated that the N.T.S.C. system is not ideal. It is designed to fit a particular type of tube, which is not a really sound basis for choosing a system which must remain valid for many years to come.

Finally, several colleagues must be thanked for many helpful discussions during the preparation of this article, particularly I. J. P. James and E. J. Gargini.



A TELESCOPIC AERIAL, which can be raised hydraulically to a height of 55ft giving a radio range of about 30 miles under normal conditions, is fitted to this second mobile office recently brought into service by the Automobile Association. It is being used for the co-ordination of the A.A.'s road weather reporting service.

Single-Pentode Flip-Flop, devised by T. E. Ivall, operates by using the valve as a pair of triodes connected in series—the bottom one acting as a cathode follower. A negative trigger pulse applied to the suppressor drives the anode, and hence the control grid, positive. Current is drawn through the screen grid, the cathode also goes positive, and, by virtue of the increased current through the cathode resistor, the suppressor is driven even more negative. A rapid cumulative action takes place, cutting off the top "triode" and rendering the bottom one fully conducting. A positivegoing trigger pulse reverses the action. Because of the complete 360° phase shift round the loop the



circuit can be operated as a sinusoidal oscillator with, say, an LC circuit in place of the anode load, or a simple RC combination in place of the potential divider connecting anode to grid. A valve with a short suppressor grid base (such as the old EF50) is the most suitable type.

Bilingual Television Transmission, with two different sound accompaniments to the picture, may be a desirable thing for some countries. A method of achieving this technically, devised by the French firm Radio-Technique, has been reported by the European Broadcasting Union. The two audio signals are arranged to amplitude modulate two interlaced trains of pulses trans-mitted in the normal sound channel on the time-division multiplex principle The time spacing of the pulses in each train is equal to the duration of a line scan, and the trains are phase locked to the line sync pulses. At the receiving end the method of separation is to apply a gating signal to one of the soundchannel i.f. valves in such a way that only the pulses carrying the wanted sound accompaniment are allowed through. The gating signal is generated by shock exciting an oscillatory circuit (tuned to the line frequency) from the line scanning circuits and then limiting the resultant complex waveform to provide the required gating voltage levels. By simply reversing the excitation leads either one or the



other of the sound accompaniments is gated as required. It is claimed that a very simple and inexpensive "decoder" unit, using no valves, is all that is required to change an ordinary television set into a "bilingual" receiver.

Binary Coded Scales are now being used on electronically controlled machine tools and other equipment where it is necessary to measure a mechanical displacement with great accuracy. They convert the analogue type of indication, such as a pointer against a scale, into a series of digits which can be read off quickly so that the task of the human operator is greatly eased. The circular scale in the sketch (made by Hilger & Watts) is for indicating angular displacement in binary digital form. The binary scale is more economical in displacement detection than other scales (e.g., 0 to 9 in decimal re-



quires ten elements compared with four in binary) and because there are only "0" and "1" elements involved it is easy to devise electrical pick-off systems for feeding the information into digital computers or other electronic data processing equipment. As well as the pure binary code, it is possible to use binary-coded decimal and other special arrangements.

"Too Old At—?" in our September, 1956, issue certainly revealed some interesting facts about people's hearing, but it didn't tell you at what stage of development or decay you have to be to hear electrical signals direct without an acoustic transducer. According to a report in the October, 1956, issue of Proc. I.R.E., engineers have experienced an audible response when standing six

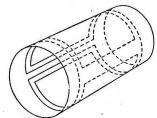
feet away from the horn of a radar set working at 1300 Mc/s with a peak power of half a megawatt. The pulse length was 2µsec and the p.r.f. was 600 c/s. The most sensitive part of the head proved to be at the sides at a point midway between the ears and eyes and slightly above them. The sounds heard were mostly high-frequency components without much of the 600-c/s fundamental, and people whose ear responses cut off at 5 kc/s heard them much less strongly than those who went up to 15 kc/s. A deaf man with a bone-conduction hearing-aid heard nothing at all. Unfortunately, such experiments can be dangerous because high-power microwave radiation can produce cataracts of the eye.

Aluminium-Wire Speech Coils, used in loudspeakers to reduce the mass of the vibrating system, bring with them the problem of soldering the wire ends. Wharfedale Wireless Works, in collaboration with Mullard, have solved the problem by dipping the wires into a small bath of molten solder agitated by ultrasonic energy. A cavitation effect in the solder removes the oxide film from the aluminium and effective tinning is achieved. Afterwards the tinned wires can be soldered in the ordinary way. A solder compound of 90 per cent tin and 10 per cent zinc is used in the bath at a temperature of about 230°C and the tin-ning operation takes 2-3 seconds. A recent improvement to the technique



has been a simplified method of maintaining the ultrasonic energy at the optimum frequency to ensure maximum soldering efficiency under all conditions.

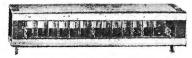
Simple Linearity Control for television line scanning, devised by J. C. MacKellar and K. E. Martin of Mullard, consists of short-circuited turns made of foil underneath the line scanning coils on the c.r. tube neck. During scanning the e.m.f. induced in the short-circuited turns is proportional to the rate-of-change of scanning flux. It causes a current to flow which, because of the R and L of the turns, changes exponentially. The flux produced by this induced current opposes the scanning flux and, if the time constant of the turns is small compared with that of the generating circuit, the waveform of the correcting flux is



more curved than that of the scanning flux. Thus the curved sections of the waveforms can be arranged to roughly match each other so that a substantially linear scan is obtained with very little decrease in amplitude. Adjustment of linearity is achieved by moving the short-circuited turns in relation to the scanning coils so that the flux linkage, and hence current induced, is varied. No "ringing" is caused, as with other types of linearity control. The sketch shows how a pair of the short-circuited turns can be constructed. They are actually joined at adjacent edges as this simplifies manufacture to producing one foil stamping instead of two. Nonlinearity can be reduced to less than 5% and the efficiency compares favourable with existing types of control.

Efficient Rectifier Cooling is the reason for the unusual "mouthorgan" construction of the new Siemens & Halske selenium h.t. rectifiers recently shown to us by R. H. Cole (Overseas). The plates are held at the edges and are arranged in groups with spaces between to give a series of "chimneys" for convection cooling. Then the edge-mounting gives a direct conduction cooling for each set of plates on to the aluminium case, which, in turn, is contact-cooled when it is held flat against a chassis by the fixing lugs. This edgemounting of the plates has a distinct advantage over the more conven-

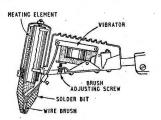
tional contact-cooled rectifiers, where the plates are stacked parallel with the case side, since it cools all of the



plates equally instead of just those at the outsides. The rectifier illustrated is a 220-V 300-mA type and measures only $3\frac{7}{8}$ in $\times 1\frac{1}{4}$ in $\times \frac{3}{4}$ in.

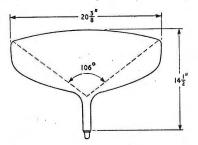
Simple Transistor Testing, using d.c. methods, may not give complete and detailed information, but it can still be useful in providing a general indication of characteristics. With this in mind, Mullard have introduced a simple instrument by which it is possible to check three of the more important junction transistor parameters, and the measurements are presented as direct meter read-For the first parameter, base-collector short-circuit current gain, advantage is taken of the approximately linear relationship between collector current and base current. This permits finite changes of current to be used to measure the parameter with an accuracy high enough for all practical purposes. Measurement is thus reduced to observing the collector current pro-duced by a convenient known base current and transcribing it into a direct meter reading of base-collector current gain. The second parameter is the d.c. collector current for zero base current. Here, since the d.c. collector current is sensibly independent of collector voltage, direct metering of this parameter can be made. Finally, for collector turnover voltage, the tester measures the collector-emitter turnover voltage for zero base current. A relatively high voltage is applied to the collector via a resistance, and the turnover voltage is read directly from the meter.

Aluminium Soft-Soldering, normally very difficult because of the tenacious oxide film which prevents "wetting" of the metal surface, should be made much easier by a special tool introduced by the Belark Tool and Stamping Company. With this, the surface of the aluminium is mechanically cleaned by a small steel-wire brush vibrating at 100c/s in the centre of the soldering bit (see illustration). Re-oxidization is



prevented by working with a pool of molten solder round the bit which excludes the air while tinning is taking place. No flux is used, but the solder has to be the special blowpipe type (80% tin and 20% zinc) with a melting point of 220°C. The soldering bit is actually heated to approximately 500°C, and this is sufficient for soldering sheet metal up to 12 s.w.g. thickness and small castings. Two pieces of aluminium which have been tinned with this vibratory tool can afterwards be joined together by orthodox soldering methods.

Wider than Wide-Angle television c.r. tube recently introduced by RCA in America makes one wonder if the flat "picture-on-the-wall" display device will really be necessary after all. The tube is a 21-inch rectangular type (21EP4) and has a diagonal deflection angle as large as 110° (or a line scan angle of 106°, as shown)! This has reduced the overall length by about $5\frac{1}{2}$ inches over comparable tubes with the same size of screen and 90° deflection angles. Another feature is a narrow neck diameter $(1\frac{1}{2}$ in), which makes



it possible to deflect the beam through the extra wide angle with only slightly more power than is required to scan a tube with a 90° angle. An electrostatic focusing system is incorporated for maintaining uniformity of focus over the whole screen. Other American tube manufacturers are following suit with the new angle.

Printed Magnetic-Cell storage device for holding binary information has recently been developed by RCA. An experimental unit has a capacity of 2,560 bits in a volume of only 2 cubic inches. It works on the magnetic-cell matrix principle described in the December, 1956, issue (p. 596) but consists of a series of thin plates of ferromagnetic material with printed conductors joining the holes. This is possible because the ferromagnetic storage medium is a ceramic-type material and therefore an insulator. The idea offers a great simplification in manufacture over the conventional matrix stores in which a complex pattern of wires has to be threaded through a great many tiny ferrite cores.

WIRELESS WORLD, FEBRUARY 1957

LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

Transistor Symbols

I HESITATE to add yet another transistor symbol to the great scrap-heap of suggestions that has accumulated already, but I think H. J. Cooke (December, 1956, issue) is working in the right direction and a simplification of his ideas might be worth considering. The following criticisms come to mind:

(1) The point transistor is practically finished, so why bother about it? Even its negative resistance characteristic, so useful for switching, can now be produced

in junction devices.

(2) The graphical distinction between p-n-p and n-p-n junction types is not really essential, since this is usually obvious from the polarity of the transistor's

power supply.

(3) Anything which involves rectangles or triangles, especially blacked-in ones, takes time to draw. (The ultimate choice of symbol will almost certainly be influenced by technicians who have to sketch it rapidly on the backs of old envelopes.) In any case, enclosed areas do not capture the interest of the eye and suggest function so effectively as the active line.

(4) Mr. Cooke's arrow for identifying the emitter of

the junction transistors suggests that this arrow is in itself the emitterwhereas in reality the emitter is the left-hand junction between n- and p-type materials indicated by the adjoining rectangles in his drawing. Here again we have an element of redundancy and possible confusion.

A logical outcome of these criti-



cisms would be a symbol something like that in the accompanying sketch. I am not particularly fond of it, but at least it is simple and has a familiar look about it without suggesting the physical construction of the point transistor.

London, S.E.5.

JAMES FRANKLIN.

Transistor Chaos

"CATHODE RAY" (November-December, issues) would find transistor symbols almost as chaotic here as in Britain. However, I would recommend to him the adoption of β for the common emitter current gain, since it is already used by the majority of American manufacturers. The symbol α can then be allowed to waste away.

Which of the resistor systems should be adopted is seemingly being decided by ease of measurement. No doubt this simplifies the work of the quality control section, but surely more emphasis should be placed on the tion, but surely more emphasis should be placed on the easing of circuit design. The great advantage of the r_0 , r_0 , r_0 , system is that external impedances can often be added in directly. With the other systems the external values must be inserted in equations which often obscure the relative importance of the various components.

M. O. FELIX.

Canadian Westinghouse Company,

Hamilton, Ontario.

"High-Quality" Demonstrations

WHILE I agree with Mr. H. Glover (October issue) that the electronic organs are of limited value as test material for high-quality sound systems, I feel that he has overlooked one of their most important features.

It is well known that those organs produce low pedal

notes. This makes them particularly useful for demonstrating the bass capabilities of a loudspeaker. In particular the records of Lenny Dee, which a number of demonstrators used, have excellent pedal notes.

A larger than usual cone area seems necessary for their reproduction, and anything less than an 18-in unit or, say, two 12-in units in a suitable enclosure would appear to be inadequate.

Crowborough.

E. R. ASLIN.

Spare Parts

THE increasing complexity of the modern television and sound broadcast receiver prompts me to ask whether the time is not fast approaching when the manufacturers should, in fairness to the buyers of their products, be willing to supply spare parts directly to the private customer in the cases where he is willing and able to fit them.

I have been employed in the radio industry for over 21 years and I well know the arguments against supplying the general public with spares. These, I submit, were perfectly valid in the days when the dealers required only a reasonable ability as repairers and when circuits were simpler. Also, in those far-off happy days the general public knew very little of radio servicing

and possibly cared less.

The position today is entirely different. There are now literally hundreds of firms employing radio "technicians "—for want of a better term—whose daily work involves far more ability than required to repair a radio

receiver of half-a-dozen valves.

On the other side, there is reputed to be a shortage of 50,000 service technicians in the retail radio trade. This simply means that having paid possibly £100 for a television receiver, the customer is unable to obtain efficient servicing if and when it breaks down.

I would, therefore, respectfully suggest, sir, that there is a case for the manufacturers to supply parts at normal prices to anyone wishing to buy, provided he quotes the serial number of his receiver. This would prevent spare-time repairers stealing too much business from the dealers, some of whom show at times a lamentable lack of enthusiasm for service work.

The owner of a television receiver would surely not order and pay for spares unless he thinks he knows what he was doing. Even if he made a mess of the job, it's

his own property he is spoiling.

It seems ludicrous to me that a man whose working day is spent among complex electronic devices should be compelled to take his five-valve teceiver to the local village radio shop to be repaired by someone who probably fills in time mending punctures!
Fakenham, Norfolk. "TECHNICIAN."

Tape Amplifier Design

I HAVE just seen R. C. Marshall's letter in your June, 1956, issue, commenting on my tape amplifier design (November, 1954, issue). The changes suggested by Mr. Marshall certainly represent a sound approach to the design of this particular feedback amplifier, although it may be mentioned that the record amplifier was found

to be absolutely stable even with the original values. It is possible to replace the triode outfit valve V_2 in the recording amplifier by a pentode. Using a large screen decoupling condenser, the amplifier oscillates with the recording head disconnected, but is stable if the recording head is connected.

tively small screen decoupling condenser is used, the amplifier will be unconditionally stable. In the latter case, V₂ operates at low frequencies as a triode, which eliminates the cause of instability. At high frequencies, in particular at the bias frequency, the screen is effectively bypassed, and the high internal impedance is preserved. At low frequencies, the additional feed-back owing to the unbypassed screen resistor is negligible and has practically no influence on the frequency

response of the recording amplifier. In my own equipment I am using a 6SH7 with a screen resistor of $1M\Omega$ and a condenser of $0.025\mu F$ between screen and cathode. R_1 , R_2 and R_{12} remain

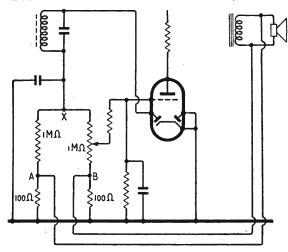
unchanged. Tel Aviv, Israel.

ARIEH F. FISCHMANN.

"Economy in Receiver Design"

THE objections raised by G. N. E. Pasch (your October, 1956, issue) to the injection of a negative feedback voltage to the grid of the first audio amplifier via the diode-load-cum-volume-control can be completely overcome, without otherwise rearranging the circuit, if a simple bridge network is used. The accompanying circuit diagram shows the arrangement; component values are suggested, but these are, of course, fairly arbitrary.

The feedback voltage being developed across AB, point X will, of course, be at earth potential so far as feedback voltage is concerned and the operation of the diode is not disturbed. On the other hand, the maxi-



mum voltage fed back to the triode grid when the volume control is at its lowest point is reduced by 6 dB in this arrangement, but this is unlikely to be a serious disadvantage in a simple receiver of this type.

London, N.10.

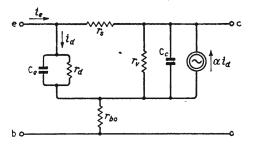
IAN LESLIE.

"Transistor R.F. Amplifiers"

IN his article in the October issue, D. D. Jones presents low- and high-frequency equivalent circuits related in a way which we feel is misleading. The low frequency value of r_e is not necessarily given by 25/I_e. This is true only if the low-frequency base resistance rb and the highfrequency base resistance r_{bo} are equal; in practice r_b is usually significantly higher than r_{bo} , so that with $I_c = \text{ImA}$, the low-frequency r_c may range down to the order of 2 ohms.

An equivalent circuit* from which Figs. 1 and 2 in the article may be derived is shown in the accompanying diagram At high frequencies r. may for all practical

*R L Pritchard, Electry Network Representation of Transistors—a Survey." I.R.E. Transactions, Vol. CT-3, No. 1, pages 5-21, March, 1956.



purposes be neglected and the circuit reduces to the high-frequency form given by Jones. To obtain his low-frequency T circuit, however, the parameter r_s cannot be neglected in the transformation. The resulting low-frequency r_e is not in general equal to r_a (=25/I_e).

A further point relates to the choice of d.c. operating conditions for the transistor. Jones mentions the reduction of Cc resulting from an increased Vc; a further benefit is that $f\alpha$ is increased.

Imperial College, London, S.W.7.

A. R. BOOTHROYD. A. R. MOLOZZI. D. F. PAGE.

Supply Voltage

I WONDER what the chairman of the South Eastern Electricity Board would do if his garage charged him for 100- and supplied him with 75-octane petrol?

Would he say, "Ah, well, the quantity is correct and the calorific value the same, so I will pay"? I somehow

But already my voltage has been as low as 199 instead of 230, and winter hasn't started yet. Seldom is my television picture quality worth watching between 7 and 9 p.m.

Last February it dropped to 177 volts and all the mains engineer can tell me is that they can't find room

for a sub-station anywhere. Banstead, Surrey.

A. R. TURPIN.

But What About "Agenda"? [Ed.]

"FREE GRID," whom I have long regarded as a firm and sure upholder of the classical basis of our language, seems to be losing his grip. Not once but twice in the January instalment of his incisive comments he treats "data" as singular. It is hardly surprising that in these days of very optional Latin this solecism should be increasingly common in technical circles, but one expects better of "Free Grid." He must surely know that " $g_m = 3.6$ " and " $\mu = 47$ " are valve data, and " $r_a = 13$ " is valve datum? Or will he soon be writing about "a phenomena" or "a cherubim"? If he lets the humanities down again I shall have to consider writing the "Second Thoughts on Love Theory" (large negative values) he so much dreads.

While on this tack may I also reproach—though more gently—"Diallist" (who, if I'm not mistaken, is another classical scholar) for referring to the present type of British TV as "monochrome"? White, as he well knows, is as far removed from monochrome as it could be, and would more aptly be named "panchrome." At the same time I do sympathize with his reluctance to keep on using the clumsy and (if the contrast control is properly set) inaccurate expression "black-and-white." It is time we had a better word.

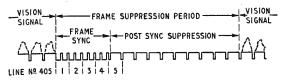
Finally, B. E. Jackson might care to note that singlesetting resistance-current-voltage-power nomograms are less uncommon than he suggests. One by W A. Barclay was published in Wireless World as long ago as February 17th, 1932, and repeated several times since; there is one in "Radio Data Charts" (5th edn.), and one in "Radio Laboratory Handbook" (6th edn.).
"CATHODE RAY."

Improved Sync Separator

Single-Pulse Frame Synchronizing for Good Interlacing

By MICHAEL P. BEDDOES,* B.Sc.(Eng.), D.I.C., A.M.I.E.E.

HE composite synchronizing signal for British television consists of alternating trains of line and frame pulses.¹ During the period in which the frame is being scanned, line pulses only are transmitted. In the small interval between the end of the frame scan and the frame flyback a group of pulses at twice line frequency is transmitted to provide a trigger for the fly-back of the frame timebase and also to maintain synchronous operation of the line timebase. In



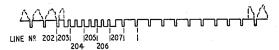


Fig. 1. B.B.C. vision waveform at the end of the frame scan; for even frames (above) and odd frames (below).

American television practice, on the other hand, the group of frame pulses is preceded and tollowed by "equalizing pulses" to facilitate accurate synchronising of the frame timebase in the receiver.

At first sight, it might appear that a single RC integrating network could produce satisfactory frame

sync pulses from the composite synchronizing signal. According to Patchett,² this is not possible because without equalizing pulses this method produces frame-sync pulses which differ slightly on odd and even frames. The slight difference makes accurate synchronizing of the frame-time-base and hence good interlacing difficult, if not entirely impossible to achieve. Various improved methods^{3, 4} have been developed for separating the frame pulses from the composite synchronizing signal. It is well known that, in certain cases, such methods can produce excellent interlacing.

However, suppose that the timebase fires slightly irregularly for odd and even frames. This would cause the initiation of flyback to vary between, for example, the first and second pulses of succeeding trains of sync pulses. Thus, although the frame lock might be considered to be satisfactory, perfect interlacing would only be possible over a strictly limited region within the locked range. According to Haantjes and Kerkhof⁵, even though a timebase has a tendency towards

irregular firing, a single, narrow and perfectly regular pulse will be conducive to the best interlacing. The single pulse frame sync separator described here is rather more elaborate than that developed by Haantjes and Kerkof.

The process of obtaining the frame-sync signal from the video signal is normally done by cascading two separators. The output of the first (the picture/sync separator) is the composite sync signal, and that of the second (the frame/sync separator) is the frame sync signal. During each frame, the corresponding sync signal may be arbitrarily divided into a group of $202\frac{1}{2}$ line pulses and one of 8 frame pulses (Fig. 1). In order to provide a frame sync signal, the frame sync separator must have an output which is produced by the sudden transition from the line pulses to frame pulses, but not vice versa. The essentials of this particular circuit are shown in Fig. 2.

The triode V_1 is driven by the composite synchronizing signal. Its anode current is completely cut off by any sync pulse; in the conduction periods between pulses the full current flows. Thus, for the purposes of explanation, the circuit of Fig. 2 can be replaced by the simpler equivalent of Fig. 3 in which a resistance R (equal to the anode resistance of the triode) in series with a switch replaces V_1 . The switch is held open during sync pulses only.

Taking the line pulses first, imagine that the switch (Fig. 3) has been closed for a considerable time, and

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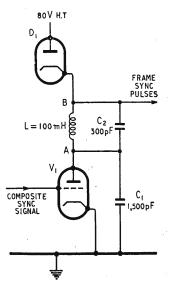


Fig. 2. Essential elements of the improved frame sync separator.

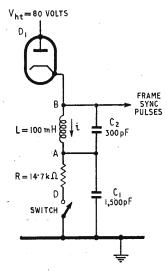


Fig. 3. Simpler version of Fig. 2, with a resistance and a switch representing the triode valve.

that the current i in the inductor has reached its steady value given by

$$i = I = \frac{V_{h+}}{R} = \frac{80}{14.7} = 5.43 \text{ mA}$$
 .. (1)

In the first line pulse, the switch is opened for $10 \mu sec.$ During this period, the damping is removed from the tuned circuit and i will decay as part of a sinusoidal oscillation whose period, T_1 (governed by L and C_1+C_2), is 83 $\mu sec.$ However, because the switch is open for considerably less than $T_1/4$, the current in the inductor cannot reach zero.

Following the line pulse, the switch is closed for 90μ sec. During this period very heavy damping is again applied to the tuned circuit and the inductive current rapidly asymptotes to the steady value I. Thus, in the period of the second line pulse, the inductive current follows a pattern similar to that outlined for the first pulse. This pattern is repeated for each line pulse in the long train of $202\frac{1}{2}$ line pulses.

Experimentally, it was easier to observe the voltage waveforms at A and B than the current in the inductance. From Fig. 4, during a line pulse v_A is elevated, implying the decay of i. In the subsequent period, the waveform of v_A indicates, also by implication, that the steady value I is reached at about

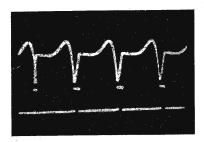


Fig. 4. Waveforms during a long train of line pulses; (above) line sync signal, (below) corresponding v_A.

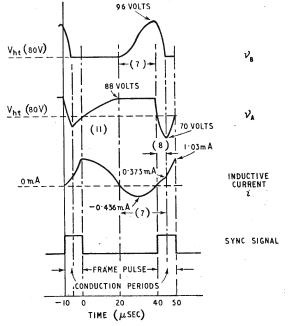


Fig. 5. Section of a long train of frame pulses. The numbers in brackets refer to the corresponding formulæ in the text.

the middle of the line scan. Nowhere within this cycle has the current in the inductor or in the diode reversed sign and therefore no output should, nor indeed does, appear at B.

Frame Pulse Pattern

Taking the frame pulses next, at the instant of the middle of the frame pulse, assume that the current i in the inductance is zero (Fig. 5) but the voltage across it, $v_A - v_B$, maintained by the charge in C_2 , is positive. The switch (Fig. 3) is open. The potential v_A will be held constant by the charge in C_1 . Although the potential v_B cannot be decreased because of the diode action, it can be increased. Therefore, during the decay of the charge in C_2 , voltage v_B is elevated as part of a sinusoidal oscillation with a period T_2 (governed by L and C_2) of $37~\mu sec$.

The switch remains open for 20 μ sec, in which period approximately half a cycle of the oscillation can take place. At the end of this period the potential difference, v_B-v_A , is now a maximum but beginning to diminish. During the ensuing conduction interval, the switch is closed for 10 μ sec, and the current through R rapidly reduces the potential v_A . Concurrently, v_B-v_A is also diminishing. In the middle of this conduction period* the potential of B falls to h.t. voltage and the diode closes. Immediately the mode of operation changes. The current in the inductance increases in the remaining 5 μ sec of the conduction period.

During the period of the next frame pulse, the switch is opened for 40 μ sec. The charge in C_1 reduces the current in the inductance and, simultaneously, the potential of A is elevated as part of a sinusoidal oscillation of period T_1 (83 μ sec). In the instant a quarter of this period (20 μ sec) later, the current in the inductance will be zero, and the voltage across it, v_A - v_B , will be positive. Thus, after this instant, a pattern follows which is similar to that already outlined. This pattern will be repeated for each frame pulse in the train of 8 frame pulses.

Experimentally, it was easier to observe the voltage waveforms $v_{\rm A}$ and $v_{\rm B}$ than the current i in the inductance. From Fig. 6, the potential $v_{\rm A}$ is elevated during the first half of the frame pulse, implying the decay of i. During the second half $v_{\rm A}$ is seen to be sensibly constant, held thus by the charge in C₁. During the subsequent conduction period $v_{\rm A}$ is seen to decay rapidly. It is elevated once again during the first half of the subsequent frame pulse.

From Fig. 7, during the first half of the frame pulse, when the inductive current and also the diode current are falling to zero, the potential $v_{\rm B}$ is seen to be constant, held thus by the action of the diode. During the second half of the frame pulse when the diode is open, this potential can be seen to increase rapidly, but it decays even more rapidly during the ensuing conduction period. In steady-state, therefore, a single narrow pulse of voltage appears during each frame pulse and at first sight it might appear that what has been described is but one more circuit for separating the trains of frame pulses 3,4 from the composite sync signal. This impression, however, is misleading, as will be shown below.

During the line pulses, the power supplied to the circuit (from considerations of the mark-to-space ratios) is 36 times that during the frame pulses.

^{*} See appendix.

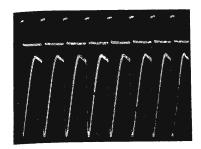


Fig. 6. Waveforms during a long train of frame pulses; (above) frame sync signal, (below) corresponding v_A.

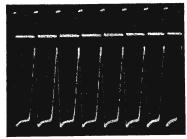


Fig. 7. Waveforms during a long train of frame pulses; (above) frame sync signal, (below) corresponding $v_{\rm B}$.

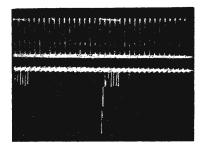


Fig. 8. Waveforms during the frame sync period; (above) composite sync signal, (below) frame sync pulse, $v_{\rm B}$.

During a succession of frame pulses, the magnitude of the voltage pulses at B is, of course, proportional to the current (approximately, because of the nonlinear elements, proportional to the square root of the power) from the valve. It is therefore to be expected that in the transition, line to frame, the first frame pulse should exceed its fellows by a factor of $\sqrt{36} = 6$. This prediction appears to be verified from Fig. 8. After the train of frame pulses, the operation of the circuit very soon† approximates to the steady-state conditions for a succession of line pulses for which there is no appreciable output at B.

Complete Separator Circuit

The circuit for a complete line and frame sync separator is given in Fig. 9. A multiple valve, V(a) and (b), performs most of the necessary operation. The pentode portion, V(a), is a classical picture/sync separator. Its input is the vision-plussync signal, while its output consists of the composite

sync signal, while its output sync signal only. This signal synchronizes the line timebase. The triode V(b) is the frame/sync separator. Its input is the composite sync signal from V(a), while its output consists of narrow pulses which are employed to synchronize the frame timebase.

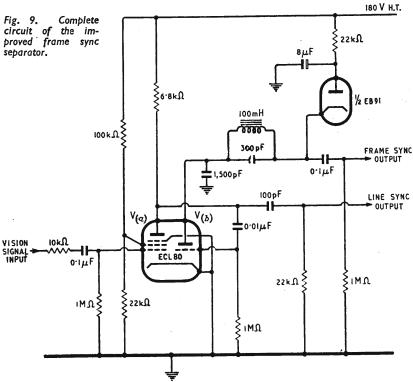
In order to limit the anode dissipation of V(b) it is fed with reduced h.t. During normal operation this valve consumes 6 mA at 80 volts: when the drive is removed the current consumption falls to 4 mA.

In discussing the performance of the circuit, one must consider the effects of interference by noise. If the frame sync-separator is either the simple RC integrator², or the more elaborate "train separator"^{3,4}, the derived sync-signal is 400 µsec or more in duration. Therefore, in order to

completely mask the frame pulse by noise, a burst of such interference must last continuously for a period which is much longer than 400 μ sec, and must coincide with the period when the sync pulses would normally occur.

For the new separator, the critical period in the composite sync signal is considerably shorter: a much shorter continuous burst of noise will therefore be sufficient to destroy the frame lock. Now the probability of obtaining a continuous burst of noise must vary inversely as its duration. Thus, in order to achieve increased precision of frame synchronism by this method, one of the prices to be paid is reduced noise immunity.

There is also the question of variations of performance between samples of the circuit manufactured by normal mass-production techniques. This frame sync circuit was employed in a production run of over a thousand television receivers with satisfactory results. The tolerances allowed on the component values were the usual $\pm 20\%$ of



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[†] All the time-constants in these circuits are very short: 22, 37 and 83 μ sec respectively

nominal, while the permitted variation in the value of the inductance L (Fig. 2) was even greater, only a minimum value of 100mH being specified. Such variations in the circuit values naturally produced observable changes in the waveforms. different receivers there was a variation in the number of minor pulses which followed the primary sync pulse; also, there appeared to be a small variation in the actual magnitude of this pulse. In spite of these variations, however, the resulting frame lock remained very precise, and the range of excellent interlace appeared to be substantially independent of the setting of the frame hold adjustment, provided of course that operation was within the synchronous

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APPENDIX

TWO numerical analyses will be made, one for the line pulses and the other for the frame pulses. In each analysis, a particular set of initial conditions will be assumed for a particular instant in the cycle; this set will then be shown to be repeated at an instant one period later.

For the line pulses, imagine that just previous to a line pulse the current i in the inductance (Fig. 3) has reached the steady value I (formula 1). During the line pulse, the switch (Fig. 3) is held open for 10 μ sec and the charges in the capacitances C_1 and C_2 will drive a current into the inductance given by the familiar oscillatory equation

$$i = I \cos \omega_1 t$$
 (

where ω_1 is the resonant angular frequency of L in shunt with the parallel combination of C_1 and C_2 and I=5.43

mA. The period
$$T_1$$
, $\left(=\frac{\omega_1}{2\pi}\right)$ is 83 μ sec. Thus, although

the current in the inductance is reduced, by the end of the $10-\mu$ sec line pulse it is not zero and consequently current still flows in the same direction in the diode (Fig. 3). Therefore, the potential at B must remain at h.t., held there by the diode action. The potential at A during this time is given by

$$v_{A} = v_{B} + \text{voltage of A relative to B}$$

$$= V_{h} - L \frac{di}{dt}$$

$$= V_{h,t} + \omega_{1} L I \sin \omega_{1} t \qquad .. \qquad (3)$$

From (3), v_A is elevated 12 volts above h.t. at the instant at the end of the line pulse. After the pulse is over, the switch is again closed. A representation of this part of the cycle may be obtained from Fig. 10, which shows the equivalent circuit. At t = 0— (i.e., just before the switch is closed), a charge equivalent to 12 volts is maintained by the capacitances C_1 and C_2 . At t=0, however, a voltage step of -80 volts is applied suddenly at point D. If e is the Laplace Transform of the voltage between

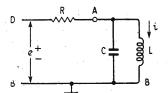


Fig. 10. Equivalent circuit for Fig 3 when the diode is conducting. Here $C=C_1+C_2=1800~\mathrm{pF}$ (from Fig. 3), while $L=100~\mathrm{mH}$ and $R = 14.7 \, k_x$.

B and D, and i is the transform of the current in the inductance, then it is easy to show6 that

$$i = \frac{e}{p^2 LCR + pL + R}$$

Mathematically, the initial 12 volts can easily be accounted for by the following device. Consider that for all time 12 volts is applied between B and D but that a step of -92 volts is applied suddenly superimposed at t=0. Then, for t>0, the -92 volts and the steady +12 volts add to produce the same effect as applying the -80 volts step, while, for t < 0, the effect is that produced by +12volts alone. Thus the current in the inductance is given

$$i = \frac{92}{p[p^2LCR + pL + R]} - \frac{12}{pR}$$
 (4)

the first main term being the contribution from the - 92 volts step, while the second term is the contribution from the steady + 12 volts. The inversion of (4) is readily performed using standard Laplace Transform techniques,7 whence

where t is in microseconds from the closing of the switch and *i* is in milliamps. Because of the heavy damping, the transient terms in (5) decay extremely rapidly. For example, after 90 μ sec (the time for the line scan) the net current is 5.43 + 0.023 = 5.47 mA, where the value 0.023 is the contribution from transient terms. Thus, the initial conditions assumed one period ago have substantially been repeated and the reasoning has completed a full circle.

On even frames, a conduction pulse of 90-µsec duration precedes each train of frame pulses. On odd frames, however, the corresponding conduction pulse is only 40 μsec, and the inductive current immediately before the frame signal is obtained by substituting t = 40 in (5). This gives i = 5.35 mA. Now, the current in the inductance immediately before the frame pulses can be regarded as proportional to the height of the first frame pulse. Therefore, the slight variation of this current (5.47 and 5.35 mA) ought to produce a corresponding variation in the magnitude of the frame pulse. Such variation was indeed observed but the small amount of this effect (2.5% computed) would probably not affect the quality of interlace. This variation could be reduced still further by using a valve with a lower anode resistance (Fig. 2).

For the frame pulses, at an instant in the middle of a

frame pulse (Fig. 5) assume the following.

(1) The current in the inductance is zero.

(2) The potential at A is elevated 8 volts above h.t. (3) The potential at B is still at h.t., but the diode current is zero.

The voltage difference across the inductance is only (Continued on page 87)

[†] Values of certain pass inductances exceeded twice nominal.

maintained by the charge in C_2 . Because the potential of A (Fig. 3) is fixed by the charge in C_1 , the potential of B, or v_B , will increase rapidly as part of an oscillation whose period T_2 is governed only by L and C_2 . Here $T_2=37$ $\mu sec.$ Thus, if time t is measured in microseconds from the commencement of the frame pulse

$$v_{\rm B} = (v_{h.t.} + 8) - \cos \frac{2\pi}{T_2} [t - 20]$$
 .. (6)

where the first main term is the voltage at A and the second term is the voltage of B relative to A.

In a low-loss oscillatory LC circuit, the peak kinetic energy in the inductance very nearly equals the peak potential energy stored in the capacitance; i.e., $\frac{1}{2}LI_1^2 = \frac{1}{2}C_2V^2$ where V is the peak voltage across the capacitance and I_1 is the peak current in the inductance. Thus, the peak current which will flow into the inductance is given by

 $I_1 = 8\sqrt{\frac{L}{C_2}} = 0.436$ mA, and therefore the current in the inductance is given by

$$i = -I_1 \sin \frac{2\pi}{T_2} [t - 20]$$

= $-0.436 \sin \frac{2\pi}{37} [t - 20]$.. (7)

From (6), it is clear that v_B can complete rather more than half of the cycle at T_2 in the 20 μ sec remaining in the frame pulse; and at the end of this (t = 40), B will be elevated about 16 volts above h.t.; while, from (7) the inductive current will be approximately zero.

During the 10 μ sec of the next conduction period, the switch is closed. Because there is no diode conduction, the potential of A can fall exponentially:

$$v_{\rm A} = 88 \exp \left[\left(\frac{t - 40}{T_3} \right) \right] \dots \qquad (8)$$

where T_3 is the time-constant formed by R and C_2 and has the value of 22 μsec . During this time also $v_{\rm B} = v_{\rm A} + \text{voltage of B relative to A}$

$$= 88 \exp -\left[\frac{t - 40}{22}\right] + 8\left[1 - \cos\frac{2\pi}{37}[t - 20]\right] ... (9)$$

The instant that $v_B = V_{h.t.}$, the diode again conducts and

the mode of operation changes. From (9) this instant occurs at t = 45; i.e., 5 µsec after the start of the conduction period; while from (7) the corresponding inductive current is + 0.373 mA. Also at this instant

$$v_{A} = 88 \exp - \left[\frac{45 - 40}{22} \right] = 70 \text{ volts} \dots (10)$$

In the 5 μ sec remaining in the conduction pulse the diode is closed, and consequently the inductive current will be given by formula (5), though with the initial constants appropriately altered:

$$i = 5.45 - \frac{70}{14.7} \left[0.265 \sin 0.716 (t - 45) + \cos 0.716 (t - 45) \right] \exp \left[-0.0198 (t - 45) \right]$$

At the end of the conduction period (t = 50) $i = 5.45 - 4.76[0.265\sin 16.35^{\circ} + \cos 16.35^{\circ}] \times$ $\exp(-0.0794)$

 $= 1.03 \, \text{mA}$

In the next frame pulse, the switch is again opened. The current in the inductance and the voltage at A are respectively given by (2) and (3) with I reduced to 1.03 mA.

Thus
$$i = 1.03\cos\frac{2\pi}{83}(t - 50)$$

and ... (11)
 $v_{\rm A} = V_{h.t.} + \frac{2\pi}{83}L1.03\sin\frac{2\pi}{83}(t - 50) \times 10^{-3}$

where t is once again measured from the beginning of the previous frame pulse and L = 100 mH.

At t = 70, from (11) $i \approx 0$ and $v_A = 88$ volts. These conditions obtain at the middle of the frame pulse and therefore the reasoning has completed the full circle

During the train of frame pulses, it has been shown that the current in the inductance immediately before a frame pulse is 1.03 mA, whereas the corresponding current immediately before the first frame pulse is of the order of 5.4 mA. Thus, the first frame pulse should exceed its fellows by a factor of 5.4. This in fact seems to be supported experimentally and closely follows the ratio derived by energy considerations alone in the main text.

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Mathematics for Electronics with Applications by H. M. Nodelman and F. W. Smith. The mathematical processes useful in electronic engineering are discussed in theory and their application to specific problems is illustrated. Sections are devoted to the use of dimensions in checking formulæ, the solution of networks by determinants and matrices, the applications of series in non-linear circuits (e.g., intermodulation testing), methods of transient analysis, and the use of Boolean algebra in the analysis of switching circuits. Pp. 391; Figs. 131. Price 52s 6d. McGraw Hill Publishing Co., Ltd., 95, Farringdon Street, London, E.C.4.

The Application of Phase-coherent Detection and Correlation Methods to Room Acoustics: B.B.C. Engineering Division Monograph No. 9 by C. L. S. Gilford, M.Sc., F.Inst.P., A.M.I.E.E. and M. W. Greenway, M.Sc. (Eng.), D.I.C., A.C.G.I., Grad.I.E.E. Describes modifications of the pulsed gliding tone technique of displaying the transient response of rooms, designed to reveal more clearly the "singularities" responsible for coloration of the sound. Pp. 14; Figs. 14. Price 5s. B.B.C. Publications, 35, Marylebone High Street, London, W.1.

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The Morse Code for Radio Amateurs by Margaret Mills. Advice on learning the code together with a graded series of exercises designed to give fluency up to a speed of 12 words per minute or more, and a glossary of amateur abbreviations. Pp. 18. Price 1s. Radio Society of Great Britain, New Ruskin House. Little Russell Street, London, W.C.1.

Wideband V.H.F. Convertor

PREPARING FOR THE INTERNATIONAL GEOPHYSICAL YEAR

By G. P. ANDERSON, A.M.I.E.E.*

A S readers of Wireless World are aware, we are approaching a maximum in the 11-year cycle of sunspot numbers, and present indications are that this maximum is going to be higher than ever previously recorded, at least during the time that it has been possible to correlate activity on the sun with its effect on radio propagation on the earth. To the average enthusiast, the most interesting and easily observable effects are to be found on the frequencies ranging from 20 to 60 Mc/s or thereabouts, and this article describes apparatus that will enable the reader to extend the tuning range of a shortwave receiver to include this band.

During the last period of "sunspot maximum", signals from all over the world were being heard on frequencies up to about 50 Mc/s; many trans-Atlantic contacts were made by amateurs on frequencies of the order of 50 Mc/s, but attempts on frequencies only a few megacycles higher (56 Mc/s) did not meet with the same success. During last winter American signals were being heard regularly on frequencies up to 36 Mc/s, and sometimes higher, with very simple indoor aerials, and, of course, the 28-Mc/s amateur band was open to the world almost daily.

Apart from possible ionospheric paths, a fair indication of tropospheric propagation conditions on v.h.f. can be obtained by monitoring the B.B.C.

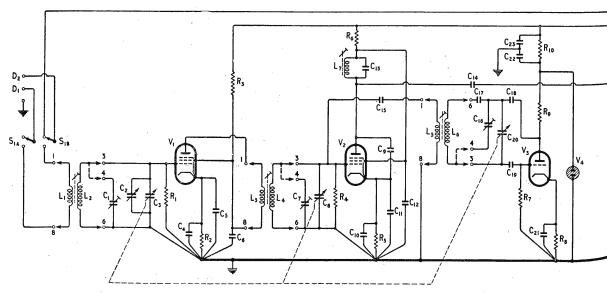
television channels, and a French TV station on about 41 Mc/s will serve as a useful sign that the 145-Mc/s amateur band is open in that direction. Final plans for the International Geophysical Year have not yet been published, but it is possible that some interesting signals may be radiated on v.h.f. in connection with this event.

Design of V.H.F. Convertors.—Due to the strength of signals when propagation conditions are favourable, fairly simple apparatus is often quite satisfactory, and a t.r.f. or "straight" receiver using modern valves could be used. However, a superheterodyne convertor used in conjunction with a shortwave receiver, or a broadcast receiver with a shortwave range, will produce much more satisfactory results. Such a convertor changes the frequency of an incoming signal on, say, 50 Mc/s, to a lower frequency, that can conveniently be about 5 Mc/s, and this "converted" or "translated" signal is then passed to a receiver tuned to 5 Mc/s.

In order to include the frequencies likely to be of most interest, the tuning range of the convertor should extend up to at least 60 Mc/s; the lower limit can conveniently be arranged to overlap the upper frequency of the existing receiver, thus, in effect, extending its tuning range. Although it is possible to use a switched system of coil changing, the writer preferred to use plug-in coils to minimize switching and other losses.

The simplest method of changing the frequency

Fig. 1. Theoretical circuit of the 5-valve convertor. The power supply is built on the same chassis.



^{*} Amateur Radio Station G2QY.

of the incoming signal is by means of a single valve of the heptode, triode-hexode, or similar type, in which one part of the valve serves as an oscillator and the voltage produced is injected into the electron stream of the other section, which functions as a mixer. Whilst such a unit is perfectly capable of giving a good performance, it suffers from the severe drawback that a strong signal is radiated from the oscillator. This may fall in the television bands and cause interference on local TV sets.

We can minimize this trouble, and at the same time obtain a useful improvement in performance, by inserting an r.f. amplifier between the aerial and the mixer; still further isolation of the oscillator voltages may be obtained by using separate valves for the oscillator and the mixer, and using loose coupling between them for oscillator voltage injection. A further refinement is a stage of amplification at the translated frequency, thereby isolating the mixer from the effects of any changes in the main receiver, as well as giving some additional amplification. For convenience of operation all stages should be ganged, and it is useful to have an aerial trimmer and a gain control available on the front panel of the convertor.

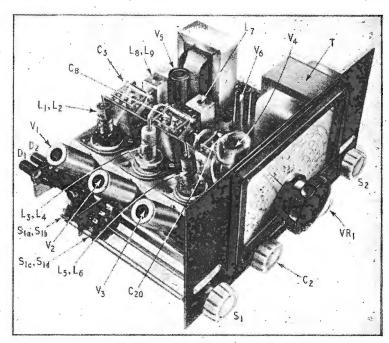
The unit shown incorporates the features dis-

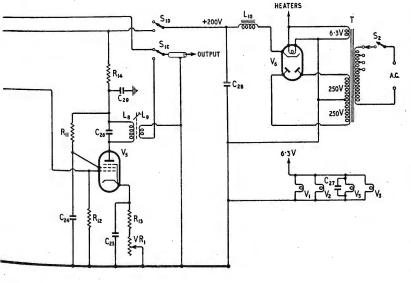
This view of the 5-valve convertor shows the inclined valves, plug-in coils, tuning capacitors and other recognizable components.

LIST OF COMPONENTS

Capacitors C_1, C_7, C_{16} 20 pF trimmer 10 pF variable 50 pF variable 470 pF C_2 $\overset{\tilde{C}_{3}}{C_{3}}, C_{8}, C_{20} \\
C_{4}, C_{6}, C_{10}, \\
C_{17}, C_{21},$ $C_{22}^{17}, C_{23}^{21}, C_{22}, C_{23}^{23}$ $C_{5}, C_{11}, C_{12}, C_{24}, C_{25}, C_{27}, C_{29}$ 1000 pF (Hi-k) 10 pF C_{13}, C_{18}, C_{26} 100 pF 47 pF 4.7 pF C_{14}, C_{19} C_{15} C_{28} $16 \,\mu\text{F}$, 350V, wkg. electrolytic.

(Unless otherwise specified, capacitors can be silvered-mica or ceramic.)





 VR_1 V_1, V_2, V_5 10 kΩ 1W, variable 6AK5 6C4 VR105/30 6X4 4-pole 2-way Yaxleytype switch "on-off" Mains switch L_1-L_6 See Table 1 50 turns No. 32 s.w.g. L_7 L_8 50 turns No. 32 s.w.g. L₉ 10 turns, ditto. (L₉ is wound over "earthy" end of L₈.) Insulation, enamel. L₇, L₈ and L₉ wound on 0.3in diam formers, Neosid Type 5000/6E; dust iron core Drg 500; top plate Drg 5001. John Dale Screening Can Type DTV2. L₁₀ 20 H 40 mA choke 250-0-250V, 40 mA; 6.3V, 1.3A.

Chassis $7\frac{3}{4}$ in x $7\frac{3}{4}$ in x $1\frac{1}{2}$ in deep. **Panel** 10in x $5\frac{3}{4}$ in deep.

cussed here, and includes simple voltage stabilization of the h.t. supply to the local oscillator. The power supply for the convertor is included in the unit. Plug-in coils are used, the range from 15 to 85 Mc/s being covered with four sets of three coils. The tuning capacitors are ganged and an aerial trimming control is provided on the front panel to compensate for the loading effect of different aerials. The trimmer also simplifies the "tracking" problem in a three-circuit tuner of this type. The other controls are: i.f. gain; convertor in/out switch, for changing the aerial over to the main receiver, and a mains on/off switch.

The circuit is shown in Fig. 1, and it will be seen that 6AK5 valves are used for the r.f. amplifier, mixer and i.f. pre-amplifier. A 6C4 valve serves for the oscillator. The h.t. voltage to the latter is stabilized by a VR105/30, the value of R₁₀ depending on the voltage of the h.t. line. The value shown is suitable for an h.t. voltage of about 200. The circuit is fairly orthodox, but the need for R1 and R₄ may not at first be apparent; they provide a d.c. path to earth for the grids of their respective valves when the coils are removed for changing the range. The capacitor C, across the mixer valve was not found to be necessary in the model shown, but from experience the writer has had with other v.h.f. convertors, it may be needed to prevent oscillation at the mixer output frequency. The power consumption of the convertor is 1.3A, at 6.3V, and 40 mA at 200V.

The main points to watch when constructing apparatus for the higher frequencies have frequently been stressed; summarized they amount to this; keep all leads short and the wiring rigid. The convertor shown is constructed with the valves inclined relative to the rest of the chassis in order to facilitate short wiring. A copper chassis should be used and the earth returns of each stage brought to one point, as indicated in the circuit diagram. Separate tuning capacitors are used and mechanically coupled, but electrically insulated from each other. Both sides of C₂₀ are "live" to r.f. and consequently this capacitor must be insulated from the chassis. It is advisable to wire the heaters with screened lead, bonded at intervals to the chassis, and unless the lead connecting the convertor to the main receiver is reasonably short, it also should be screened. A good slow-motion dial is desirable, although a certain amount of fine tuning can be carried out on the main set.

The coils are wound on $\frac{1}{2}$ -in diameter formers, provided with $\frac{3}{8}$ -in dust cores for adjustment of inductance. The particular type specified in the schedule is made of polystyrene, and use may be made of the low softening temperature of this material to secure the turns of wire during the winding process. The writer found it convenient,

after soldering one end of the wire in Pin 3, to bend it at right angles just above the flange of the former (see Fig. 2), and to hold the wire so that it is resting against the body of the former. If now the wire is heated by carefully applying a soldering iron to it at a point near to the body of the former as shown, it will be found that after a very short time the wire begins to sink into the polystyrene. The iron should now be removed, and the wire and former held in position for a few seconds to allow the poly-styrene to harden. Winding the coil may then pro-ceed, keeping the wire taut, separating the turns, and, if felt necessary, repeating the heating process at intervals. After the required number of turns has been put on, the last turn should be secured in the same way as the first and the end inserted into Pin 6 and soldered. The other winding may be added in the same way, and to complete the coil a coating of polystyrene varnish should be applied, which will effectively "embalm" the windings, and prevent any risk of turns moving during handling. If formers of clear polystyrene are used they may be colour coded by applying a coat of paint to the inside of the former.

The coils used in the prototype unit plug into standard octal sockets, which may be colour coded to correspond to the appropriate coils. These sockets, like the valveholders, should be of high-grade ceramic or similar material, with good v.h.f. properties. Details of the windings are given in Table I, from which it will be seen that certain ranges are "padded" with extra capacitance in order to spread the tuning over the more populated frequencies; with one exception these capacitors are fitted in the wiring of the convertor (at C_1 , C_2 and C_{16}) and are brought into circuit by joining tags 3 and 4 of the appropriate coils. It may also be observed that Band B includes both the 21- and the 28-Mc/s amateur bands.

Provided the layout and coil data are followed fairly closely, no difficulty should be encountered in aligning the convertor to provide an output signal for a set tuned to 5 Mc/s. The first step is to adjust the i.f. pre-amplifier to this frequency, and this may be done by setting "i.f. gain" to maximum (with the convertor switched on and connected to the receiver), and adjusting the dust cores in L₇, L₈ and L₉ for maximum "noise" in the loudspeaker. The ranges covered by the oscillator may then be set using a signal generator or a grid-dip oscillator, and remembering to adjust the inductance at the low frequency end of each range, and the capacitance trimmer at the high frequency end, if necessary. In the prototype unit one value of capacitance served for most ranges, but due to differences in wiring it may be necessary to use different values for each range, in which case the capacitors could be mounted on each coil; this was done in the case

TABLE I. COIL WINDING DETAILS

Range	Coverage	L, Turns	L, Turns	L, Turns	L, Turns	L, Turns	L. Turns	Notes
Α	15-22 Mc/s	2	8‡	2	8‡	2	71/3	Pins 3 and 4 joined.
В	21-30 Mc/s	- 1	5∤	2	5 <u>‡</u>	2	41	Pins 3 and 4 joined.
С	30-53 Mc/s	1	23	2	41	2	3	Le has 13pF across it.
D	40-85 Mc/s	1	3	2	3 1	2	31⁄2	

All are wound with 22 s.w.g. tinned copper wire, on 1-in diam slug-tuned formers (Denco Maxi-Q, 6-pin octal based). Enamelled wire may be used if preferred. Turns are spaced by the diameter of Individual pairs of coils for each range (i.e. L, and

the wire. Individual pairs of coils for each range (i.e. L_1 and L_2 ; L_3 and L_4 ; L_5 and L_6) are wound on the same former, end to end with about $\frac{1}{12}$ in spacing between them.

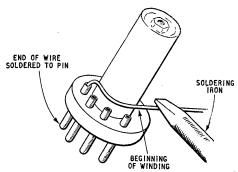


Fig. 2. The end turns on the polystyrene coils can be fixed in position by applying a hot soldering iron as shown here.

of L₈ (Range C), where an additional 13 pF was needed. The value of the padder capacitor C₁₇ appears to be a good compromise and gives satisfactory tracking over all bands. However, some experiments may be useful in individual cases to obtain optimum results.

Having adjusted the oscillator range, the signal frequency circuits may be aligned, again adjusting the inductors for maximum signal near the low frequency end of each range, and the capacitors at the h.f. end. (This subject has been dealt with thoroughly in a number of publications and the reader is referred to them if he should require further information.)

Switch S₁ enables the convertor to be put in or out of circuit without having to manually change-over plugs, and in the "out" position the aerial is connected to the main receiver, and the convertor is disconnected.

Aerials.—The unit may be used with more or less any length of wire as an aerial, and if a single "long wire" aerial is used, it should be connected to terminal D_2 of the convertor, D_1 being strapped over to the earth terminal E. Care should be taken to make these connections correctly, so that the aerial is connected to the output socket when S_1 is set to "out."

On the higher frequencies, however, greatly improved results can be obtained by using resonant aerials; that is, aerials the lengths of which are chosen to tune to the particular frequencies in which one is interested. A convenient length consists of a half-wavelength of wire, that is, one 5 metres long for reception on 10 metres, and although such an aerial connected as described for a "long wire" will give good results, its efficiency is severely limited by the fact that, being only a short length of wire, most of it will be screened by being inside a room. It is preferable, therefore, to place the half-wavelength of wire as high and as clear of buildings as possible, and connect it to the receiver by a transmission line, that will convey signals from the aerial with little loss, and at the same time pick up minimum signal itself, thus reducing the effects of man-made static. Much information has been published on aerial design, but for the purpose of this article it will be sufficient to give the formula for determining the length of a half-wave aerial; this is

$$l \text{ (feet)} = \frac{462}{f \text{ (Mc/s)}}$$

If the wire is cut at the centre and an insulator

inserted to separate the two halves, and a length of suitable feeder used to connect it to the receiver, it can be erected well away from the house. Balanced feeder of the correct characteristic impedance of 75 or 80 ohms is available, but an alternative may be made from electric light "flex," preferably plastic insulated, which has approximately the same impedance. It is, however, more "lossy," and is liable to be even more so in damp weather. In either case the ends of the two conductors forming the feeder should be connected to D_1 and D_2 , the strap from D_1 to E not being required; terminal E may be connected to earth. Coaxial feeder of the type used for connecting TV aerials and receivers would be satisfactory also, the centre conductor being connected to D_2 , and the sheath to D_1 .

The half-wave aerial favours reception of signals from directions at right angles to its length, and this factor can be made use of if reception is required mainly from a particular direction; but for general (or omni-directional) coverage the aerial may be erected in a vertical position. As a half-wavelength at 28 Mc/s is about 16 ft and at 50 Mc/s only about 9 ft it is quite a practical proposition to erect such an aerial. The design of more elaborate aerials and arrays is beyond the scope of this article, and in any case such aerials are unnecessary for the listener who only wants to hear "what's going on" on the very high frequencies.

CLUB NEWS

Birmingham.—A demonstration of high-quality sound reproduction will be given to members of the Slade Radio Society by a representative of Whiteley Electrical Radio Company on February 15th. At the first meeting in February (1st) G. A. Swinnerton (G6AS) will speak on operating in the DX bands. In addition to the fortnightly meetings held at 7.45 at Church House, High Street, Erdington, there are instructional and constructional classes every Tuesday and Wednesday evening, and the club station (G3JBN) is available every day for use by members. Sec.: C. N. Smart, 110, Woolmore Road, Erdington, Birmingham, 23.

Bradford.—The subjects to be covered by speakers at the meetings of the Bradford Amateur Radio Society on February 12th and 26th are, respectively, oscilloscopes (by G. F. Craven), and transformers (by P. Howarth). Meetings are held at 7.30 at Cambridge House, 66, Little Horton Lane. The fortnightly meetings are preceded by half-hour morse classes. Sec.: F. J. Davies (G3KSS), 39, Pullan Avenue, Eccleshill, Bradford, 2.

Bury.—The February meeting of the Bury Radio Society will be held at 8.0 on the 12th at the George Hotel, Kay Gardens, when H. A. Rothwell (G6QT) will demonstrate an all-transistor broadcast receiver. Sec.: L. Robinson, 56, Avondale Avenue, Bury.

Leicester.—Dekatrons and other counter tubes will be covered by M. H. Kind (G3GXZ) in his lecture to the Leicester Radio Society on February 11th at 140, Highcross Street. Sec.: J. Tranmer, 4, Grocot Road, Evington, Leicester.

Newbury.—J. H. Etheridge, of Cinema-Television, will speak about the flying-spot particle resolver at a meeting of the Newbury and District Amateur Radio Society on February 22nd. Meetings are held at 7.30 at Elliott's Canteen, West Street. Particulars are obtainable from 83, Newtown Road, Newbury.

Sidcup.—The next meeting of the Cray Valley Radio Club will be held at the Station Hotel, Sidcup, on January 22nd at 8.0. C. Usher (G2CCD) will give a talk entitled "Antennas for the amateur." Sec.: S. W. Coursey (G3JJC), 49, Dulverton Road, New Eltham, London, S.E.9.

Wellingborough.—On February 28th G. Abrams will address members of the Wellingborough and District Radio and Television Society on "Basic audio amplifiers." Arrangements are being made for a Mullard lecture and film on valve manufacture on the 21st. Meetings are held each Thursday at 7.30 at Silver Street Club Room. Sec.: P. E. B. Butler, 84, Wellingborough Road, Rushden.

NEGATIVE RESISTANCE

The Answers to Last Month's Ouestions

By "CATHODE RAY"

A FRENCH reviewer* has remarked that it is my custom to take supposedly simple things and show how obscure they are; and then, when the fog has become really dense, to wave a magic wand and clear it away. That may be all very well when the fog obeys, but last month it tended to thicken continuously, necessitating a postponement of the trick. In case you haven't meanwhile found your own way out or lost interest, the attempt will now be made, with apologies for any inconvenience.

The problem, you may remember, concerned negative resistance—as provided, for example, by dynatron or transistor—in circuit with positive resistance. Both kinds of resistance can be represented as voltage/current graphs. Fig. 1 shows the

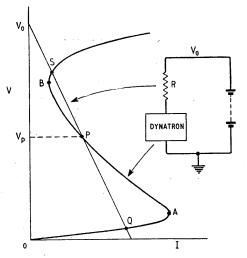


Fig. 1. Voltage/current graphs of the two resistive elements in the circuit shown. The curved line QAPBS represents, to the scales beginning at the normal origin O, the resistance of the dynatron, which is negative between A and B. The straight line V_oSPQ represents the wholly positive resistance R, to an inverted voltage scale beginning at V_o.

characteristic curve of a dynatron; between A and B its resistance is negative, indicated by the line sloping the "wrong" way—current decreasing with rise in voltage. The straight line represents the linear positive resistance R; and to see how the total voltage V₀ is shared between it and the dynatron we plot it to an inverted voltage scale, beginning at V₀ as its zero and increasing positively downwards. So the downward slope of this line indicates positive resistance all the way. The amount of resistance appears as steepness of slope. The only points on the diagram corresponding to equal current through both resistances are those which are common to both lines.

With two positive resistances only one such point is possible, but a peculiarity of negative resistance is that the same total voltage can sometimes be divided between the two resistances in more than one way, with different currents; for example, P, Q and S in Fig. 1.

We found by experiment that we could set the circuit to point P by tying down the junction between positive and negative resistances to a tapping $V_{\rm p}$ on the voltage source, but that this point was unstable, because immediately the tapping was disconnected the voltage flipped to Q or S and stayed there. If, however, R was lower, as in Fig. 2, only one situation was possible.

Again by experiment we found that the negative resistance provided by a point-contact transistor with common base bias has a different kind of V/I curve, as in Fig. 3; but that in spite of the fact that at and near P the relative slopes of the positive and negative resistance are the same as in Fig. 2, this circuit turns out to be unstable and flips to Q or S just like Fig. 1. So our theory, based on Figs. 1 and 2, that the circuit is unstable when R is greater than the negative resistance and stable when it is less, broke down, and we were faced with Thomas Roddam's poser-how can the circuit, set to P, know whether it is stable or unstable without going to see which way the negative resistance bends? And how can it do that if it is stable? And how can it do it anyway, seeing that it can't move from P without there being more than one current or voltage in the same place at the same time, which is impossible?

We had been given a possible clue to the last of these questions by Mr. Roddam, when he pointed out that no real circuit is free from reactance. In practice

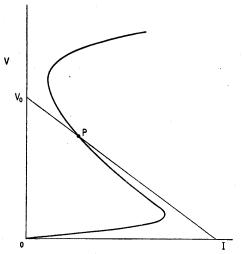


Fig. 2. As Fig. 1, except that R is less than the negative resistance of the dynatron.

^{*}F.A. Toute La Radio, November 1955. p. 391.

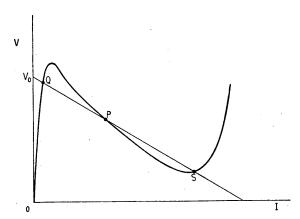


Fig. 3. Here the curved line represents the input resistance of a transistor circuit (Fig. 6).

there is always some stray capacitance across the dynatron in Fig. 1, and the current from this as its voltage changes between P and S or Q is in the right direction to bridge the current gap between the two lines during this change, and the amount of current required controls the speed of the change. We demonstrated this by making the capacitance large, slowing the process down so much that we could easily follow it on milliammeters.

That particular problem seemed to have been disposed of very neatly until we applied it to the transistor circuit. Because the two lines above and below P in Fig. 3 are on opposite sides of one another as compared with Fig. 1, the capacitance current is only in the right direction to bridge the gap between the two lines when the voltage is moving towards P, which therefore ought to be a stable point. But experiment showed that even with as much as $300\,\mu\text{F}$ across the transistor it was most

definitely unstable!

Meanwhile we had got no nearer the answer to Roddam's question; and our shunt capacitance theory, demonstrable though it was in connection with Fig. 1, led us into the most frightful dilemma with Fig. 4, which illustrates the transistor in series with a high resistance and shunted by a capacitance. Near P it is identical with Fig. 1, so presumably the slightest displacement causes capacitance current that displaces it more, and so on, just as happened with the dynatron; but the current gap, instead of closing up at Q or S, widens continuously, causing the capacitance voltage to change faster and faster; but the negative resistance characteristic makes that impossible, for the voltage change actually reverses! We tried it, and found that the current and voltage oscillate continuously to and fro between the bends. But how does it manage to get into reverse there, contrary to our current-gap theory? And how is P in Fig. 4 unstable in spite of there being only one intersection, as in the stable Fig. 2?

Two Kinds of Resistance

It was with all these awkward questions unanswered that I was callous enough to leave you last time. In the meanwhile I have thought up some answers, which I hope will have been worth waiting for

It is difficult to know where to start, but first of all let us tackle the question why P in Fig. 2 represents

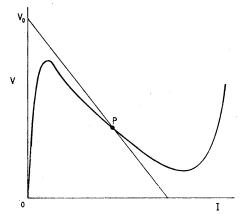


Fig. 4. As Fig. 3, except that R is greater than the negative input resistance of the transistor circuit.

a stable condition whereas in Fig. 3—which in the region of P is absolutely the same—it is unstable. That was the original Roddam problem.

Because diagrams of various kinds are such valuable mental aids, there is a danger of relying on them too completely, overlooking their limitations. Our ideas (if you will pardon my assuming that you, too, were taken in last month were heavily based on voltage/current graphs, in which negative resistance appears as a slope downward from left to right. From this point of view there is no difference whatsoever between a portion of negative-resistance characteristic plotted from a dynatron and one from a transistor, provided their slopes are both the same. Yet they behave quite differently in practice. Since negative resistance is a particular relationship between current and voltage, and it is that relationship which is expressed by the graph, can two identical pieces of graph possibly represent different

kinds of negative resistance?

The difficulty in seeing a difference between two resistances that appear identical may perhaps be that we are too much under the spell of Ohm's law. If we increase the e.m.f. applied to a 500-ohm resistor by 10 volts we know that the current through it will rise by 20 milliamps. We also know that if we increase the current through it by 20 milliamps the p.d. across it will rise by 10 volts. Cause and effect are interchangeable. It is the same with pressure and rate of water flow through a pipe. But if a fireman uses the water to knock a man down, cause and effect are not interchangeable. A man falling down doesn't necessarily cause water to gush out of a hose. And because the relationship between current and voltage in a valve or transistor can be graphed in the same way as for a resistor, it does not entitle us to assume that the same reversible cause-and-effect holds good. We have already seen that although the passing of I_p milliamps through the device illustrated in Fig. 5 necessarily causes the p.d. across it to be V_p volts, applying V_p volts to it doesn't necessarily drive I_p milliamps through it. It might be Iq. Or Is. In this case it is only by regarding current as the cause and voltage the effect that we can get an unambiguous result. This kind of negative resistance is therefore called current-controlled. Conversely, the dynatron is a voltage-controlled negative resistance.

The persistent inquirer will object that distinguishing them by these names doesn't properly

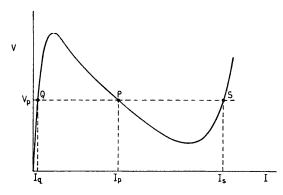


Fig. 5. This is the characteristic curve of a currentcontrolled negative resistance, because the value of current definitely decides the potential drop, whereas the reverse is not true.

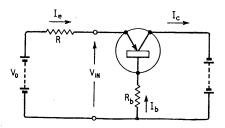


Fig. 6. Diagram of transistor circuit used to experiment with current-controlled negative resistance.

answer the question, for in order to name them we had to take a look at the bends in the curve, whereas the actual circuit appears to know the difference while it is still on the straight near P. So we shall have to go into the matter more closely.

Treating the transistor characteristic in Figs. 3 and 4 in the same way as for ordinary resistance, we would conclude that raising the input current from point P was bound to make the voltage fall, and reducing the voltage was bound to make the current rise. But let us look at the circuit represented by this graph: Fig. 6. Here, I, and Vin are the current and voltage defining the transistor input resistance. Suppose we raise the input current Ie, either by raising an already large Vo through an R sufficiently large for any changes in the potential of the emitter to have only minor influence on I, or by reducing R. These alternative procedures are represented in Fig. 4 by raising V_{\circ} , keeping the slope of the R line the same, or by decreasing the slope of the R line, keeping Vo the same. In either case, the diagram shows that P is shifted downwards, indicating the rise in current we aimed at, and also a fall in voltage. The same result can be predicted from Fig. 6 by a knowledge of point-contact transistor action. The rise in I_e, which, without I_c, would cause the voltage drop across R_b to increase, so increasing V_n, actually causes a bigger rise in I_o, which passes through R_b the opposite way, so the net result is a fall in V_{in}. The circuit behaves as the graph predicts.

Now let us reduce $V_{i.}$ directly, by reducing $V_{i.}$ through a comparatively low R as in Fig. 3. Fig. 3 predicts a resulting rise in $I_{o.}$ And so it would be if cause and effect were reversible, as with ordinary resistances. But let us consider it with Fig. 6. The first effect of a small reduction in V_{in} is to

reduce the voltage available for driving current from emitter to base. So I_e falls. That immediately causes a bigger fall in I_e, which makes the base less negative; that is, the base potential rises to meet the falling emitter potential, further reducing I_e. So the slightest reduction in V_{in} causes I_e to fall with a trigger action, the rate of cut-off being limited only by the time taken by the electrons and holes inside the transistor to do their stuff, as studied by us in the last September issue. The process is only saved from going on for ever by the top bend, where the transistor input resistance becomes positive and a stable position Q is taken up. If the trigger had been pulled in the other direction—as it can be now, by raising V_e until the hump above Q has been cleared—it flies to the alternative stable point S.

Effect Follows Cause

We see, then, that an actual transistor circuit reacts to a voltage change near P in exactly the opposite way to that predicted by the V/I diagram. This is because it is a current-controlled negative resistance, so is affected by the applied voltage change only in so far as that changes the input current, which in turn changes the input voltage in the opposite direction. But it can't work in the reverse order, any more than the man falling down can make water come out of the hose. It is true that if, when we lowered Vo, some playful demon had intervened for a microsecond or two, increasing I, just in advance of us, our action would have sustained this result after the demonic influence had disappeared, for the increase in Io would by then have made the base move considerably more negative than the emitter, so increasing the voltage between them and hence also Ie and Ic, in accordance with Fig. 3. Presumably also a shunt capacitance would then have acted always in such a way as to turn any excursions back towards P, so making it a point of stability. But science is based on the assumption that demons, if any, do not interchange cause and effect in this way; and Fig. 3 is invalid for control by voltage. Hence the complete and instant failure of every experiment I could devise to hold the circuit at P.

Incidentally, my fanciful closing remark last month (about my having to see this month's issue to get the answers) was really a rather subtle clue, because the only reason for its fancifulness was its reversal of cause and effect.

The problem of the gap between the two lines when moving from P to Q or S does not arise, because the extremely brief period of the journey is occupied by the internal electronic processes I mentioned, during which time the transistor has no static characteristic curve that one can draw on a V/I diagram.

In the same way it can be seen that because the negative resistance of a dynatron is voltage controlled it is unstable with a resistance greater than the negative slope, as at P in Fig. 1, whence there would be a flip to Q or S even if all shunt capacitance could be completely removed. The fact that the capacitance explanation is also true for this particular arrangement is just one of those awkward coincidences that make it so easy to confuse two different issues and draw fallacious conclusions. Fig. 1 is just Fig. 3 with current and voltage interchanged; it is,

(Continued on page 95)

in fact, the dual* of Fig. 3, and since the dual of shunt capacitance is series inductance we would expect series inductance to have the stabilizing influence we sought from shunt capacitance in the transistor circuit. The diagram with which we "proved" the stability of P in Fig. 3 also serves for inductance in Fig. 1 if dualized by interchanging current and voltage (Fig. 7), for the current-bridging effect of the changing voltage across the shunt capacitance becomes a voltage-bridging effect due to a changing rate of current through the series inductance. But, as with the transistor, this stabilizing effect is of academic interest only, for it is overridden by the impossibility of effect preceding cause.

In short, the answer to our original question (How does the circuit know whether it is unstable, as in Figs. 1 and 3, or stable, as in Figs. 2 and 4, without departure from P to see whether the negative resistance curve closes in to a Q and S or opens out?) is that a given negative slope around P in a V/I graph can represent either of two different kinds negative resistance—current-controlled voltage-controlled-and since this nature of the negative resistance in the actual circuit determines which is cause and which is effect, and effect is bound to follow cause, the stability or instability is determined. We have already studied Fig. 3 at length and seen why P is unstable. In Fig. 2, where the diagram at P is identical, the situation represented is nevertheless different, because a slight lowering of voltage is a direct cause in its own right, which increases the current, which admittedly drops somewhat further the voltage available for the valve and so extends the original action, but (with the relatively low positive resistance represented in Fig. 2) only to a limited degree. An infinitesimal displacement from P is sufficient for the nature of the negative resistance to be tested and for the circuit to flip or stay put as the case may be.

But what, you may ask, if a current-controlled negative resistance had a characteristic curve as in Fig. 2—or just continued negative all the way, without bends—and flipped madly on at goodness knows how many volts and milliamps per microsecond, trying in vain to find a stable point that did exist?

The continuous negative resistance can be ruled out for a start, because negative resistance is not a thing that can come from a passive component like a resistor; by its nature it delivers more power than it receives, and since there is no such thing as "perpetual motion" the power must come from somewhere, such as a battery, which of course is limited; and usually the device through which this power is administered—the valve or transistoris even more limited, and sooner or later in both directions its resistance must revert to positive —hence the bends. The bends being granted, they must be such as to make possible more than one current at a given voltage, or more than one voltage for a given current, as in Fig. 2. If the former, voltage cannot be the controlling interest, else there would be no ambiguity; the negative resistance must be current-controlled. Vice versa for Fig. 2. So there is no fear of our nightmare coming true.

If you suppose that by this time the now celebrated Roddam question has been completely and

satisfactorily answered from all possible angles and we can go home, may I remind you of two little things. The first is that although some of us jibbed at Roddam's own answer as being not entirely obvious to the dim-witted, we never questioned its correctness. His answer, if you remember, was based on the existence of circuit reactance. Now our following up of this clue was rather unfortunate, because it led us quite a long way up the garden, and it was only when we dismissed reactance altogether as the determining factor in the problem that we got somewhere. Was Roddam wrong, then? Or are we?

Oscillations or Flips?

It seems (I hope so, anyway) that we are both right as far as we go. The discrepancy is removed if we recognize two kinds of instability: the kind we have been considering, which we might call flip instability; and the kind Roddam was considering, which we might call oscillation instability. Our kind is the inability of a circuit with negative-resistance to stay at the middle point of three; for example, P in Figs. 1 and 3. The reason is the fact that effect comes after cause, not before; a fact that over-rides anything that reactance can do. The gap between the negative and positive resistance characteristic curves needs no filling, because during the flip the negative-resistance characteristic is in a state of flux and doesn't really exist as a static curve.

The second thing to recall is that in attempting to set the circuit an impossible and absurd task we stumbled across oscillation instability—with capacitance across the transistor and a high resistance in series, as represented by Fig. 4. (Incidentally, the same thing happens with a dynatron and low resistance (Fig. 2) if there is inductance in series—as the dualists among us would expect.) Theory seemed to indicate that the nightmare "irresistible-force-and-immovable-object" situation would take charge, the speed of flip increasing to infinity and even beyond. But the thing oscillated to and fro,

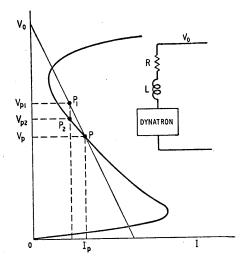


Fig. 7. A series inductance L makes possible a gap $V_{\rm p1}$ — $V_{\rm p2}$ between the voltages across dynatron and resistance R, provided that the current through all three circuit elements is changing in the direction away from P.

^{*}Wireless World, April 1952; also Second Thoughts on Radio Theory. Chap. 35.

and we haven't yet found any explanation for the reversal at each bend.

When I put an oscilloscope across R_b in the transistor circuit it showed that the speed of flip was indeed very great. One could tell not only by the fact that during it the trace was as near vertical as one could see, indicating an immeasurably small time period, but this vertical trace was almost invisible, showing that the ray was moving at very high speed. However, the speed of current change could never actually reach infinity, for at least two The first is the already-mentioned finite speed of electronic action inside the transistor, which renders the negative-resistance curve invalid at very high rates of change. The second is that every circuit has some inductance, and however small it was it would set up an infinitely large voltage if the current changed infinitely fast. So the current from the shunt capacitance not only causes an increasing horizontal separation of the two working points P₁ and P₂ but also a vertical separation, representing the voltage of self-induction. One can therefore visualize the two points starting off side by side; then one rising above the other, overtaking it horizontally, and driving the other in the opposite direction, where it follows.

In a transistor circuit without added inductance, in which oscillation is of a sawtooth and pulse (or relaxation) type, my guess is that the electronic time delay is likely to be the larger influence. If inductance is added in series with the capacitance, one gets an acceptor circuit which is in effect a low-impedance dynamic load in parallel with R, swinging it round from the Fig. 4 position to the flipunstable position (Fig. 3). But now it is not merely flip-unstable, for being a dynamic or oscillatory load the inductance reverses the motion smoothly at the end of each half-cycle, and oscillation is more

nearly sinusoidal.

The behaviour of the dynatron, with series inductance to convert the stable Fig. 4 into oscillation instability, is the dual of what I have just described. But the electronic action is much swifter in the dynatron, and the inductance is invariably shunted by at least some self-capacitance, making a rejector circuit, so my guess is that capacitance is here the main reversing influence. The effect of the high-impedance oscillatory circuit in series with the comparatively low R in Fig. 2 is to swing the load line up to the Fig. 1 position.

So in the end we find that, out of Figs. 1-4, 1 and 3 are the unstable ones, and 2 and 4 the stable. This is on the understanding that 2 and 4 do not have hidden about them enough reactance to bring the load line at some frequency or other into the

If R in Figs. 1 and 3 is a simple positive resistance it is effective at all frequencies down to zero, and there is no reversing agent, so it just flips once to a stable point, Q or S. Supposing there could be no reactance at all the speed of slip would be governed entirely by the speed of electronic processes in the negative-resistance device. During this short period its characteristic curve would be changing in such a way as to eliminate any gap between itself and the resistance line. Reactance slows the flip by filling a current or voltage gap.

If \overline{R} is the dynamic resistance of a resonant circuit it depends on frequency. The simplest sort, having one lumped inductance L and one capacitance C,

in addition to inevitable resistance, is over-all resistive at only one frequency. If L and C are in series, the result is an acceptor circuit, which at resonance is a low resistance, and therefore likely to form an unstable combination with a currentcontrolled negative resistance. If indeed its resistance is lower than the negative resistance, continuous oscillation will take place at the frequency of resonance, for the interchange of current and voltage twice during each cycle causes reversal at the bends by a combination of the current-gap effect studied last month and the voltage-gap effect illustrated in Fig. 7. Because current is the thing common to both L and C in a series circuit, it is easy to remember (even apart from Fig. 3) that it is the appropriate kind for oscillation with current-controlled negative

Conversely, the parallel-resonant or rejector circuit, in which the same voltage is across both L and C, is the one for voltage-controlled negative resistance; it provides the high dynamic resistance which Fig. 1 shows is needed for instability. In spite of the appearance of Fig. 1, it is effectively in parallel with the negative-resistance device; the battery Vois there simply as bias, and at the frequency of oscillation can be regarded as a short-circuit.

The dynatron and the point transistor with base bias resistance are not much more than technical curiosities, which would hardly justify the time we have given to them. But they typify the much more used devices in which negative-resistance is created by positive feedback. So the same principles apply to all kinds of positive-feedback circuits, which would be quite something even if we didn't remember that most negative feedback is positive at some frequencies.

Chromium Nitride Resistors

RESISTORS of small dimensions and very high stability are sometimes required for special types of equipment, and investigations at the Battelle Institute in America' have shown that it is possible to meet these requirements with resistors constructed of films of chromium nitride (Cr-N), or of chromium-titanium nitride (Cr-Ti-N). Films of these materials are deposited on ceramic bases by the vacuum-evaporation method and resistors of up to several megohms can be made having temperature coefficients of resistance less than 0.01% per degree C.

Nitriding is carried out at temperatures ranging from about 950°C to 1250°C and the films so treated exhibit wide ranges of temperature coefficient and of resistance per square.² Under certain conditions the temperature coefficient changes from positive to negative and the investigations point to the possibility of producing a wide range of resistors with near-zero temperature coefficients. The stability is greatly improved by mounting the resistors in sealed glass containers.

¹Reported in the (American) Journal of the Electrochemical Society Vol. 102, No. 2. February, 1955.

² The normal expression for resistance of a conductor is given by:—

 $R = \frac{\rho \times l}{t \times w}$

where R=resistance in chms, ρ =resistivity in ohm-cm, l=length, t=thickness and w=width in cm. Resistance of films of constant t and l=w is known as "resistance per square."

FEBRUARY MEETINGS

LONDON

4th. I.E.E.—"The importance of research in hearing and seeing to the research in hearing and seeing to the future of telecommunication engineering" by Dr. E. C. Cherry at 5.30 at Savoy Place, W.C.2.

8th. Television Society.—"Scatter

8th. Television Society.—"Scatter propagation and its application to television" by Dr. J. A. Saxton (D.S.I.R. Radio Research Station) at 7.0 at 164 Shaftesbury Avenue, W.C.2.

12th. I.E.E.—"The ultimate performance of the single-trace high-speed oscillograph" and "The design and performance of a new single-transient oscillograph with very high writing speed" by M. E. Haine and M. W. Jervis at 5.30 at Savoy Place, W.C.2.

13th. Radar Association.—"Automation: computer-controlled machine tools

tion: computer-controlled machine tools for small quantity production" by D. T. N. Williamson (Ferranti) at 7.30 at the Anatomy Theatre, University College, Gower Street, W.C.1.

lege, Gower Street, W.C.1.

14th. British Kinematograph Society.

"Photo-electronic aids to photography" by Professor J. D. McGee at 7.15 at the Royal Society of Arts, John Adam Street, Adelphi, W.C.2.

15th. B.S.R.A.—"Some recent developments in amplifiers" by F. Langford-Smith at 7.15 at the Royal Society of Arts, John Adam Street, W.C.2.

15th. Institute of Navigation.—
"Navigation and traffic control over the

W.C.2.

15th. Institute of Navigation.—
"Navigation and traffic control over the
north Atlantic" by D. O. Fraser at 5.15
at the Royal Geographical Society, 1,
Kensington Gore, S.W.7.

15th. Physical Society and Institute
of Physics.—"Acoustics in non-destructive testing" by Dr. R. W. B. Stephens
at 6.30 at 47 Belgrave Square, S.W.1.

20th. I.E.E.—"The Stereosonic recording and reproducing system" by
H. A. M. Clark, Dr. G. F. Dutton and
P. B. Vanderlyn at 5.30 at Northampton Polytechnic, St. John Street, E.C.1.
21st. Television Society.—"The
design of oscilloscopes for television
laboratory work" by O. H. Davie
(Cossor) at 7.0 at 164 Shaftesbury
Avenue, W.C.2.

26th. I.E.E.—Discussion on "The
analysis of waveforms" opened by A.
Cooper and D. A. Drew at 5.30 at Savey

analysis of waveforms" opened by A. Cooper and D. A. Drew at 5.30 at Savoy

Place, W.C.2.
27th. Brit. I.R.E.—"Some applications of nucleonics in medicine" by E. W. Pulsford and N. Veall at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

BELFAST

12th. I.E.E.—"Electronics and automation" by Dr. H. A. Thomas at 6.30 at the Engineering Department, Queens University.

BRISTOL

25th. I.E.E.—"Germanium and silicon power rectifiers" by T. H. Kinman, G. A. Carrick, R. G. Hibberd and A. J. Blundell at 6.0 at Electricity House, Colston Avenue.

27th. Institution of Production Engineers.—"Automation" by S. P. Woodley at 7 15 in the Small Lecture Theatre, University Engineering Laboratory, University Walk.

CAMBRIDGE

19th. I.E.E.—"Electronics and automation" by Dr. H. A. Thomas at 8.0 at Cavendish Laboratory, Free School

20th. Brit. I.R.E.—"Radioactivity and its measurement" by E. W. Pulsford at 6.30 at the Cardiff College of Technology and Commerce, Cathays Park.

CHESTER

27th. Society of Instrument Technology.—"Computer-controlled machine tools" by G. S. Kermack (Ferranti) at 7.0 in the Board Room of Chester and District Hospital Committee, 5 Kings Buildings, King Street.

EDINBURGH

22nd. Brit. I.R.E.—" The field evaluation trials of electronic equipment" by H. Holmes and "The electronic manipulation of digits applied to statis-tics" by J. Kyles at 7.0 at the Depart-ment of Natural Philosophy, University of Edinburgh.

GLASGOW

14th. Brit. I.R.E.—"The earth satellite project" by P. H. Tanner at 7.0 at the Institution of Engineers and Shipbuilders, 39, Elmbank Crescent.

28th. Institution of Production Engineers.—"Electronic control" by J. A. Stokes at 7.30 at the Ruston Club.

LIVERPOOL

14th. Brit, I.R.E.—"Radioactivity and its measurement" by E. W. Pulsford at 7.0 at the Chamber of Commerce, 1 Old Hall Street.

MANCHESTER

MANCHESTER
7th. Brit. I.R.E.—"Electronics in medicine" by R. F. Farr at 6.30 at the Reynolds Hall, College of Technology, Sackville Street.
12th. Society of Instrument Technology.—"Computing technique applied to measurement and control" by J. Wills at 7.30 at the College of Technology. nology.

13th. I.E.E.—" Frequency-modulated quartz oscillators for broadcasting equipment" by W. S. Mortley at 6.45 at the Engineers' Club, Albert Square.

NEWCASTLE-ON-TYNE
18th. I.E.E.—"The use of transistors in radio and television" by Dr. A. J. Biggs at 6.15 at Kings College.

I.E.E.- "The B.B.C. sound broadcasting service on very-high frequencies" by E. W. Hayes and H. Page at 6.30 at the College of Technology and Arts.

WOLVERHAMPTON

13th. Brit. I.R.E.—"An automatic system for electronic component assembly" by K. M. McKee at 6.0 at the Wolverhampton and Staffordshire Technical College, Wulfruna Street.

January Amendment

In place of the I.E.E. discussion on January 29th on the performance of d.c. amplifiers, mentioned last month, four papers will be read, including "A four papers will be read, including Abridge network for the precise measurement of direct capacitance" by A. C. Lynch, and "A simple transformer bridge for the measurement of transistor characteristics" by W. F. Lovering and D. B. Britten.



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WIRELESS WORLD, FEBRUARY 1957

RANDOM RADIATIONS

By "DIALLIST"

Guarantees

TO ME it has always seemed ridiculous that a new sound or television receiver should not be covered by a single comprehensive guarantee. As it is, there is a twelve months' setmaker's guarantee, but this does not cover the valves and cathode-ray tube for which there are separate three- and six-month guarantees. And there's another absurd point. The purchaser isn't covered at all for consequential damage caused by the failure of a particular component. Suppose, for example, that the breakdown of a capacitor damages several valves. Then the only free replacement to which you're entitled is that of the capacitor. The valves, you see, didn't blow up through any defect in their materials or manufacture. That kind of guarantee isn't worth having.

Twelve Months' Comprehensive

What I'd like to see is a comprehensive 12-month guarantee for the whole set covered by a single registration card, returnable by the purchaser to the manufacturer of the set. After all, the set manufacturers pay the piper and they can call the tune. If they stuck out for comprehensive twelve-month guarantees to them by valve and cathode-ray tube makers, they'd get them. I'm sure that the result would be increased sales. A hesitant customer is much more likely to take the plunge if he can feel that he knows exactly where he is with his set for the next twelve months. I've heard not a few people say that they won't buy a television set with its big array of valves and its cathode-ray tube until valve and tube makers show sufficient confidence in their products to guarantee them for more reasonable periods.

An Old Stager

THE OTHER day I was shown what must be one of the oldest wireless receivers still at work. Built in 1923 or 1924, it had originally five "R" valves—and what the output from its horn loudspeaker must have sounded like one shudders to think! It now has four 2-volt triodes and a power valve feeds a moving-coil loudspeaker. These are almost the

only alterations that have been made. Few repairs had been needed, I was told; but valve replacements had naturally been necessary at intervals. Still in its original home-made cabinet, it is a strange looking thing and vastly bigger than the five-valve set of today. I imagine that the sets-both sound and television—that we think so neat to-day will look just as cumbersome and as weird to those who see them in museums thirty odd years from now. Transistors will undoubtedly reduce the size of both, and flat cathode-ray tubes will slim television sets till they measure no more than five or six inches from back to front.

Direct or Projection?

THE FLAT cathode-ray tube will undoubtedly be perfected both in this country and in America. But are big tubes the best answer to the demand for big pictures? Myself, I very much doubt it. I've always thought projection the sounder and more scientific way of providing a large-sized television image. One wonders why it has not caught on better with the viewing public. It hasn't, for the number of projection sets seems to grow less at each succeeding Radio Show. Amongst its advantages are that, size for size, the projected picture is usually a good deal less liney than the directly viewed. Then there's the difference in cost when a new cathode-ray tube is needed. I believe that what has hindered the progress of the projection receiver is that the optical system needs rather skilful adjustment and that servicemen capable of carrying this out are too few and far between.

Holme Moss or Emley Moor?

A FRIEND who has recently been in Leeds tells me that he heard one or two complaints of interference by the v.h.f. sound transmissions from Holme Moss with television programmes from the same station. This, if it's a fact, is rather a curious business. The television frequencies are 48.25 Mc/s (sound) and 51.75 Mc/s (vision). Those used for v.h.f. sound are 89.3, 91.5 and 93.7 Mc/s. Further, vertical polarization is used for TV and horizontal for v.h.f. One imagines, too, that the v.h.f. carrier frequencies must have been carefully chosen to avoid any chance of their interfering with television reception. Will any reader living in those parts and who can throw light on the matter please let me know? I'd have thought myself that interference would be more likely to come from the I.T.A. Emley Moor station, working in Channel 10

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(sound 196.25 Mc/s, vision 199.75 Mc/s). Holme Moss television aerials would mostly be designed for the mid-frequency, which is 50 Mc/s. It doesn't seem unlikely that they might be excited by a 199.75 Mc/s signal at short range and with 200 kW behind it.

Seeing Into the Past

WHAT an interesting pursuit the new branch of science, radio astronomy, must be. As Professor A. C. B. Lovell pointed out in his Kelvin Lecture it was the wartime advances in electronics in general and radio in particular that gave it a start. Though it was discovered in 1932 that radio waves from outer space were reaching the earth, it was not until 1950 that the first radio telescope came into action. Since then over 2,000 centres of radiation have been tracked down and this number is sure to be enormously increased when the giant telescope at Jodrell Bank gets to work next year. What I find specially interesting is that some of these centres are nebulæ, which are all that is left of stars which once grew rapidly to abnormal brilliance and finally blew up. One of these nebulæ is the remains of such a supernova recorded in A.D. 1054. Since it is 4,000 light (or radio wave) years away the explosion must have occurred about 3000 B.C. and the radiations now reaching radio telescopes began their long journey some time before the first stone was put into place at Stonehenge.

French TV System

THE MAP in the December W.W. showing the locations of European television transmitters gives one the impression that those of the French system are rather queerly distributed. Except for Paris and Bourges all the stations seem to be near the north coast and the northern and eastern frontiers. Many important towns must be very much in fringe areas, if they get a TV service of any sort. I'm sure there's some very good reason for the odd looking distribution of the transmitters, for the French authorities undoubtedly know their business and are determined to make television a success in their country. A possible explanation is that the map only takes account of stations that were actually in use last July and that what now look like gaps will soon be filled by transmitters still under construction.*

* R.T.F. plans to have some twenty stations operating by the end of 1957.—ED.





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Crystal Jubilee

I SPEAK subject to correction but, to the best of my knowledge and belief, last year we should have celebrated the jubilee of the crystal detector which first came into use in 1906. As we all know, the crystal followed the valve which in the form of a diode rectifier was patented by Fleming on November 16th 1904. The crystal subsequently came into its own as it had a better characteristic curve than Fleming's diode and so was more sensitive.

I have little doubt that many of you veterans will tell me that I am all wrong and that you used a crystal rectifier long before 1906. What I really have in mind is its commercial use which in 1906 meant in ship and

shore stations.

It is always difficult to fix a date for the first use of a particular technique. For instance it is generally agreed that the magnetic detector started to come into commercial use in 1902 when it rapidly displaced the coherer. But this particular magnetic detector was really only the first commercial version fathered by Marconi; the one with the continuously moving iron band. Rutherford used a magnetic detector of sorts to receive signals across Cambridge in 1897.

A Bucolic Bugbear

WRITING in a journal which circulates only among members of the radio trade, a dealer calls attention to the large number of battery sets in use in country districts which have

no electric power supply.

Owing to the high cost of using dry cells continuously for L.T. supply, two-volt accumulators still flourish to a remarkable extent, and this dealer alone handles over 100 each week. I must confess that this rather surprises me as accumulators are messy things to have in the house but, worse still, have to be lugged periodically to the charging station and brought home with the week's shopping.

Surely the obvious thing to use as a substitute for a dry l.t. cell is the parent from which it sprang; namely, the ordinary wet Leclanche cell. It is true that this can be equally as messy as an accumulator but it does not have to be taken to the charging station. Whatever can be done by the small dry cell can be done very much more economically by a wet one as it only needs a new zinc electrode at very infrequent

Polarization is the bugbear of Leclanché cells, wet or dry, but the l.t. demand of a modern battery set

is so small that the depolarizer in the porous pot of the wet cell is more than ample to keep it going. Even if it were not, it would be an easy matter to rig up two cells and a change-over switch so that one could have a rest while the other was working.

Another advantage of the Leclanché is that it does not need a voltage-dropping resistance when used with 1.4-volt valves as does a 2-volt accumulator.

Teledynamics?

THE EDITOR and I have, in recent months, both written about the transmission of power by radio. No doubt it will be many years before serious consideration is given to this question. But one day power transmission by wireless will "arrive" and we ought to coin a proper portmanteau word for it. We don't want to be caught napping as we were in the case of television and have a horrible hybrid word like "dynamission" foisted on us by the lay press.

I cannot think of a good and correct word offhand; all that suggests itself is "teledynamics." While correctly derived, this word doesn't suggest the idea of radio transmission any more than "telearchics" sug-

gests radio control.

I am not too fond of any of these "tele" words in any case. The word "telephone" does not, for instance, suggest the idea of transmission by wire or by any form of electrical energy. It would be equally correct to apply the word "telephone" to a speaking tube.



Newnham Nymph

Now it is up to some of you classical boffins from Balliol to lend me a hand and justify your existence. I feel sure the "Greeks had a word for it" but it needs a lot of excavating

by some of the Bodleian boys. However, it is really women who never seem to be at a loss for words, and so perhaps I had better call on the Newnham nymphs for help in this matter.

Destaticizing

I HAVE previously denounced as mere superstition the dainty little trailing chains which many motorists used to destaticize (what a word!) their car bodies and so eliminate the travel sickness to which many people are prone. If it be other than superstition why is it that railway travellers are not immune from sickness, for few things are more firmly earthed to the rails than a railway carriage? Why, too, don't cyclists suffer from sickness for their machines are as much insulated from the road as a car?

A correspondent has, however, put forward an argument which to him seems to prove that the destaticizing chain is really effective in preventing travel sickness even in cases where the psychological effect of autosuggestion can be ruled out.

It appears that a dog owned by a colleague of my correspondent, which has always been prone to car sickness, has been completely denauseated by the fixing of one of these dangling chains. Therefore, argues my correspondent, the chain must work by physiological rather than by psychological means as the dog does not know that its sickness is caused by ambient static which the dangling chain removes.

Now whatever our views about the modus operandi of telepathy it is well known that a dog readily picks up the mood of a human being, and more especially that of its master. It is also well known that this phenomenon is due, not to telepathy, but to teleolfaction. Human and other beings exude a certain amount of perspiration even in an arctic temperature, and the odour of the perspiration varies with the emotions.

We have all heard of the "odour of sanctity" and Shakespeare speaks of the "disodour of impiety." There is, in fact, a characteristic odour associated with every emotion, and a dog, with its keen sense of smell, sorts them out. Thus a strange dog is apt to attack people who are afraid of it as they emit an odour of fear.

Obviously my correspondent's

Obviously my correspondent's colleague, who owns this particular dog, has great faith in the dangling chain and he must, therefore, exude a strong odour of faith—not necesarily akin to the odour of sanctity—which his dog readily picks up.

WIRELESS WORLD, FEBRUARY 1957







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Frequency: 4–220 Mc/s in 7 expanded bands. Calibration accuracy: $\pm 1\%$. Output: A.M. 100 mV, F.M. and sweep 3 mV. Attenuation: 0, -20, -40 db. Output impedance: 75 ohms unbalanced. Sweep frequency: 50/60 cycles. A.M. & F.M. audio modulation: 900 c/s approximately simusoidal. A.M. modulation depth, F.M. deviation and sweep band-width all controllable.

Power supply: 105/125 V. or 200/250 V. A.C., at 40/100 c/s. Consumption

Weight: 20lb. Dimensions. 17in. x 9in. x 8in. (43 x 23 x 20 cm.) Finish: Steel case in grey hammer finish with Perspex escutcheon

LIST PRICE £75.0.0.

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MY3-275
Power Triode



Mullard

The MY3-275 triode has been introduced by Mullard for audio power amplification at high levels.

It has a thoriated tungsten filament and its maximum anode dissipation is 275W. Under typical class 'AB2' conditions, two of these valves in push-pull will provide an output of more than IkW. Further data for the MY3-275 and other large output valves and details of Xenon and Mercury Vapour rectifiers for associated power supplies are readily obtainable from the address below.

Abridged Data for the MY3-275 (Services Type Numbers CV1252 and CV1619)						
V _f	14.0 V	Typical operation in-class 'AB	for two valves 2' push-pull			
· · · · · · · · · · · · · · · · · · ·	6.5 A	Va	3000 V			
Pa	275 W	I _a (o)	2x60 mA			
*g _m	8.5 mA/V	I _a (max. signal)	2x 280 mA			
*μ	16	R _{a-a}	13,800 Ω			
*ra	1900 Ω	Pout	1300 W			
*Measured at	Va=2000V, Ia=160 mA	D _{tot}	3 %			

MULLARD LIMITED, COMMUNICATIONS & INDUSTRIAL VALVE DEPARTMENT

for the closest approach to the original sound

The criterion, as always, is that the reproduced sound shall be the closest approach to the original—that the enjoyment and appreciation of music may be unimpeded. This is reflected throughout the design of the QUAD II. It is reflected, too, in the straight-forward and logical system. of control, achieved without the sacrifice of a single refinement or adjustment capable of contributing to the final objective. Send for further details and booklet.









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DE-FLUXER

-a simple, easily operated trouble-free device for depolarising the heads of tape recorders and players.

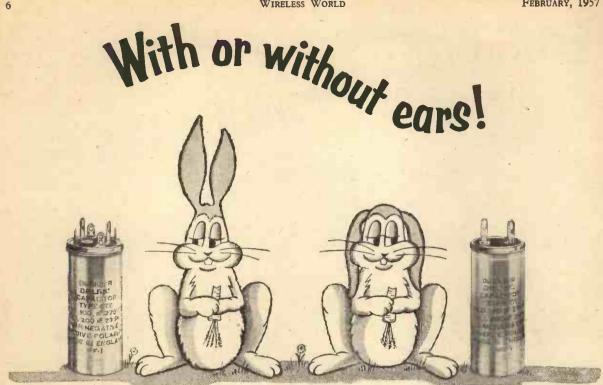
- It ensures the maximum signal/noise ratio from any tape recorder.
- rotects valuable recorded tapes from cumulative background noise and the gradual attenuation of the higher frequencies.
- * Is extremely simple to use without removing head screening cans.
- Permits selective tape erasure during editing.

The Wearite Defluxer is a "must" for all tape recorder hi-fidelity enthusiasts and for the professional recordist.

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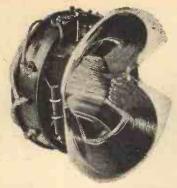
For example, Dubilier can supply you with electrolytic capacitors for television receivers made either for ear mounting* or clip mounting. In either case they are manufactured with the high ripple current sections required for this purpose. These capacitors are assembled and sealed in seamless drawn aluminium cans.

*For fixing ear mounting types, only four slots are required in the chassis. The capacitor is dropped into these slots and a slight twist of the ears secures capacitor firmly. Alternatively, a bakelite mounting plate can be supplied for use in those cases where isolation of the capacitor can from chassis is required.

Capacitance (μF)	D.C. Wkg. Voltage	Dimensions	Ripple Current (mA)
100—200	275—275	4" x 13"	700—300
50	280	3" x 13"	500
100	280	2" x 13"	550
50—100	350—280	3" x 13"	500—200
50—100	280—280	3" x 13"	450—200
200—500	350—350	4" x 2"	700
64—120	350-350	4" x 13"	500
100—200	350-280	4" x 13"	900—300
100-200	350-280	4" x 13"	700
60-100	350-350	4" x 13"	500—200
60-250	350350	4" x 1¾"	700—400
100—100	350-350	4" x 13"	550-200
100-200	350-350	4" x 13"	900—300

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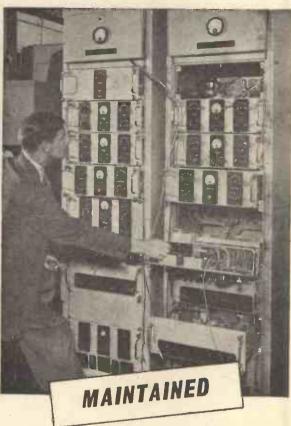


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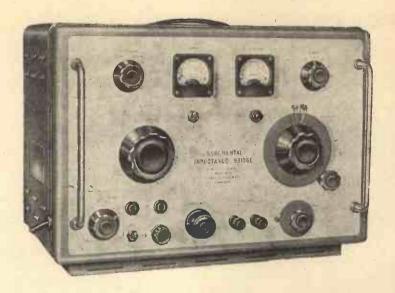
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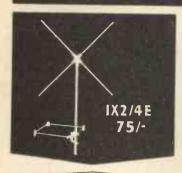
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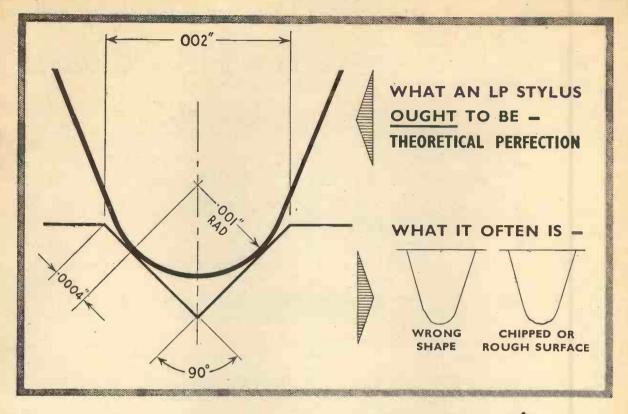


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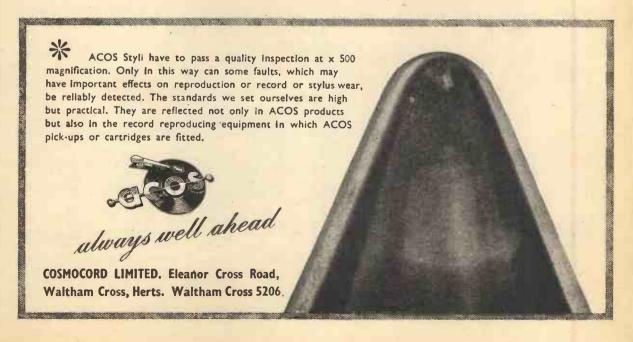
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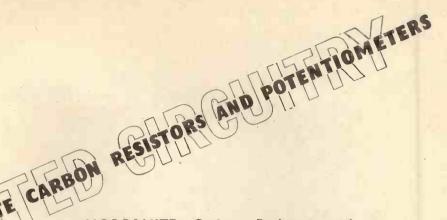
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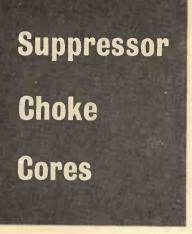
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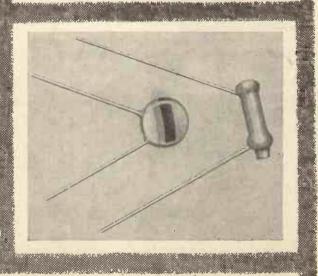
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CT10Q/2	2.2 pF		1/1½d	1/0d	CT20Q/2	200 pF	1/9d	1/7½d · ·	1/6d
CT10Q/2	2.7 pF	_	1/1½d	1/0d	CT20Q/2	240 pF	1/9d	1/7½d	1/6d
CT10Q/2	3.3 pF 3.9 pF		1/1½d	1/0d 1/0d	CT25Q/2 CT25Q/2	300 pF 330 pF	1/9d 1/9d	I /7½d I /7½d	1/6d 1/6d
CT10Q/2 CT10Q/2	5.0 pF	1/1 <u>4</u> d	1/0d	10 ² q	CT30Q/2	390 pF	1/94	1/7 ² d	1/6d
CT12Q/2	7.5 pF	1/14d	1/0d	10 ¹ d	CT35Q/2	510 pF	1/9d	í'/7≟d	1/6d
CT12Q/2	8.2 pF	1/114	1/0d	10 ¹ d					
CT12Q/2	10.0 pF	1/1½d	1/0d	10½d	HIGH	-K TUBULAR	(Tolerance	-20% +80%)
CT15Q/2	15.0 pF	I/I≟d	1/0d	10½d	List No.	Capac	itance	List I	Price
CT20Q/2	20.0 pF	l/l∮q	1/0d	I0½d	Eige 110:				,
CT20Q/2	22.0 pF	1/1½d	1/0d	10 ¹ d	CTIOK/2		0 pF	10	
CT25Q/2	30.0 pF	1/1 1 d	1/04	10 ² d	CTIOK/2		0 pF	10	
CT25Q 2	33.0 pF	1/1½d	1/0d	10 [‡] q	CT10K/2 CT10K/2		O pF	10	
CT30Q/2	39.0 pF	1/1½d	1/0d	IV2q	CTIOK/2		0 pF	10	d
Type N750				CTI5K/2		0 pF	1/9		
CT10Q/2	10 pF	1/1½d	1/0d	10½d	CT18K/2 CT20K/2		O pF	1/0	
CT10Q/2	15 pF	1/1½d	1/0d	I0∮d	C120K/2	3,00	о рг	./	ou .
CT10Q/2	20 pF	1/1½d	1/0d	l0½d	LII	GH-K DISC (To	Janania 201	0/ 1 200/)	
CT10Q/2	22 pF	1/1½d	1/04	I0½d	- Ini	GH-K DISC (IC	nerance -20	% T00%)	
CT10Q/2	24 pF	I/I½d	1/04	10 ¹ d	CD 8K/2	47	0 pF	10	†d
CT10Q/2	30 pF	1/1½d	1/04	10 ⁷ q	CD 8K/2		0 pF	. 10	id /
CT10Q/2	36 pF 39 pF	1/1½d	P0/1	10 [§] d	CD 9K/2	, 1,00	0 pF	- 10	
CT10Q/2 CT10Q/2	47 pF	1/1½d	1/0d	10 ² q	CD 9K/2		0 pF	10	
CT10Q/2	51 pF	1/1½d	1/0d	10 ¹ d	CD11K/2 CD12K/2		0 pF	10	
CT10Q/2	68 pF	1/13d	1/0d	10 d	CDI4K/2		0 pF	io	
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All prices are for speakers without transformers.

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MEASURING INSTRUMENTS & APPARATUS >



133 WEST SATH STREET NEW YORK 11, N. Y. + OREGON S-7820

3 Stock machines selected at random! Gentlemen:

January 16, 1956

We have tested the three Garrard Model 301 Turntables which the undersigned selected at random from sealed unopened cartons in your warehouse stock. These three bore the following serial numbers: 867. 937, 3019. We used a standard Model WB-301 mounting base without modification, a Leak tone arm fitted with their LP cartridge, and a complete Leak preamplifier and power amplifier, model TL/10.

Pickup and amplifier system conformed in response to the RIAA-new AES-new NARTB curve within \pm 1 db.

Standards referred to below are sections of the latest edition, National Association of Radio & Television Broadcasters Recording and Reproducing Standards.

Our conclusions are as follows:

Turntable Speed

Measurements were made in accordance with NARTB specification 1.05.01, using a stroboscope disc. In every case, speed could be adjusted to be in compliance with section 1.05, i.e. within 0.3%. In fact, it could easily be adjusted to be exactly correct.

Wow

Measurements were made at 33-1/3 rpm in accordance with NARTB specification 1.11, which calls for not over 0.20% deviation.

Garrard	Serial	No.	,	%
	867			.17
	937			.13
	3019			.12

These values substantially agreed with those given on Garrard's individual test sheets which are included with each motor.

Rumble

Measurements were made in accordance with sections 1.12 and 1.12.01, using a 10 to 250 cps band pass filter, and a VU meter for indication. Attenuation was the specified 12 db per octave above 500 cps and 6 db per octave below 10 cps. Speed was 33-1/3 rpm.

easily adjusted to exact speed!

Turntable

WOW less than NARTB specifications!

Rumble less than most professional recording turntables!

*

Mr. C. J. LeBel

President of the Audio Instrument Co., Inc.;

Chairman of one of the groups which prepared the NARTB Standards; Founding member of the Audio Engineering Society, past president; Member of the Acoustical Society of America.

Audio Instrument Co., Inc., makers of special high calibre test equipment for use in laboratory measurements.

Garrard 301 has been designed to provide the professional user and quality enthusiast with a unit supreme in its class — truly the world's finest transcription turntable for home use!

to those seeking finest results in a transcription turntable

leph <code>TESTED</code>: for performance by Audio Instrument Company, Inc., an independent laboratory.

RESULTS: Garrard Model 301 tested even better than most professional disc recording turntables...sets a new standard for transcription machines!

Read Mr. LeBel's report below



Signal to Rumble Ratio,
Using Reference Velocity of 7 cm/sec at 500 cps

This reference velocity corresponds to the NARTB value of 1.4 cm/sec at 100 cps.

Garrard Seri	al No.	DE
867		. 52
937		49
3019		49

The results shown are all better than the 35 db broadcast reproducing turntable minimum set by NARTB section 1.12. In fact they are better than most professional disc recording turntables.

Signal to Rumble Ratio,

Using Reference Velocity of 20 cm/sec at 500 cps

Garrard Serial No.	DB
867	61
937	58
3019	58

We include this second table to facilitate comparison because some turntable manufacturers have used their own non-standard reference velocity of 20 cm/sec, at an unstated frequency. If this 20 cm/sec were taken at 100 cps instead, we would add an additional 23.1 db to the figures just above. This would then show serial number 867 to be 84.1 db.

It will be seen from the above that no rumble figures are meaningful unless related to the reference velocity and the reference frequency. Furthermore, as stated in NARTB specification 1.12.01, results depend on the equalizer and pickup characteristics, as well as on the turntable itself. Thus, it is further necessary to indicate, as we have done, the components used in making the test. For example, a preamplifier with extremely poor low frequency response would appear to wipe out all rumble and lead to the erroneous conclusion that the turntable is better than it actually is. One other factor to consider is the method by which the turntable is mounted when the test is made. That is why our tests were made on an ordinary mounting base available to the consumer. to the consumer.

Very truly yours,

AUDIO INSTRUMENT COMPANY, INC.

C. J. Lekel

CJL:ds

Rumble: checked by official NARTB standard method (-35 db. min.) -52 db.!

Rumble: checked by Manufacturer A's methods -61 db.!

Rumble: checked by Manufacturer B's methods -84.1 db.!

Of greatest importance! Always consider these vital factors to evaluate any manufacturer's claim.

Now there's a Garrard for every high fidelity system











MODEL 301

RC 98/4

RC 88/4

RC 121/4

TPA 10

GARRARD ENGINEERING & MANUFACTURING CO., LTD., SWINDON, WILTS., ENGLARD

CATHODEON

Miniature Crystal Oven

Hew Release.

This miniature oven has been designed specifically for accurate temperature control of Ministry Style "D" (Cathodeon Type 2M)
Crystal Units. It is particularly recommended where the requirement is for a frequency stability better than .0002% over a wide range of ambient temperatures.



Centre tapped heater for 6.3 or 12.6 volt supplies.

Current consumption 0.73 amp. at 6.3 volt.

Temperature differential $\pm 2^{\circ}$ C within a temperature range 75°C to 85°C.

Temperature control within the ambient range—20°C to +70°C. Heating time less than 5 minutes from ambient temperature to 85°C.

Precision bi-metallic contacts with provision for On/Off indicating lamp.

Low loss Mycalon base.

Weight 1½ oz. (43 grammes).

Overall Dimensions $1\frac{1}{4}$ " x $\frac{13}{16}$ " x $2\frac{5}{8}$ " long (3.2 x 2.0 x 6.6 cm.).

CATHODEON CRYSTALS LIMITED LINTON CAMBRIDGESHIRE

TELEPHONE LINTON 223

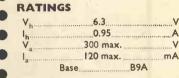


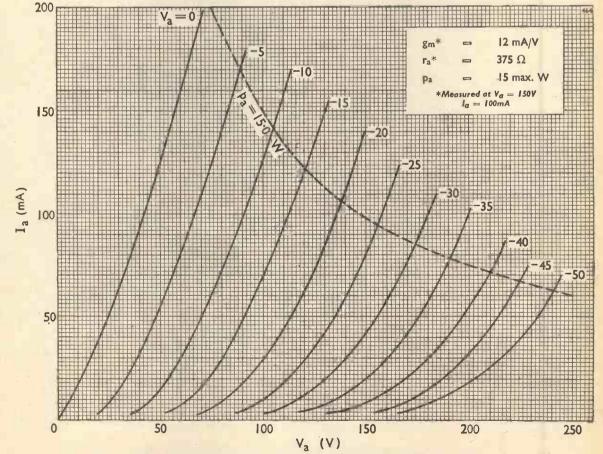


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For further details write to the G.E.C. Valve & Electronics Dept.



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The new RCA Electron Ray Tuning Indicator makes exact tuning simplicity itself.

* No interference

The FM system coupled with RCA circuitry results in exceptional signalto-noise ratio.

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87.5-108 Mc/s covers the entire international F.M. broadcasting band.

★ Great Sensitivity

2 microvolts for 20 db quietening— extends the 'fringe' 7 valves plus 2 crystal diodes and Electron Ray tuning indicator.

* High Fidelity

Wide range response within 1 db from 20-15,000 c/s for true High Fidelity reproduction.

- ★ No Matching Problems Adjustable output levels.
- * Automatic Frequency Control Ensures complete freedom from drift.
- * Power Requirements

230-390 volts D.C. at 40 milliamps H.T. supply. 6.3 volts, 2.25 amps heater supply (available from RCA New Orthophonic High Fidelity power amplifier).

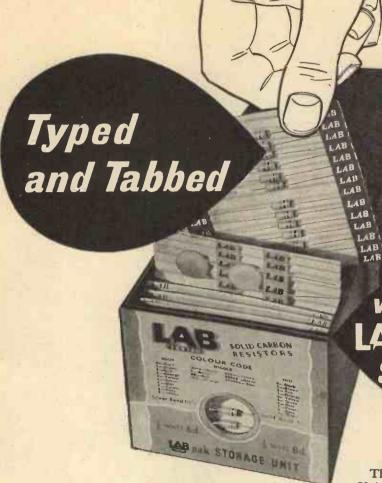


BRITAIN

(An associate Company of Radio Corporation of America)

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Tel.: Sunbury-on-Thames 3101.



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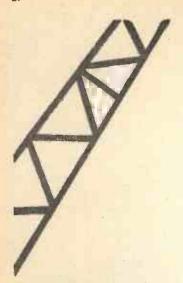
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1 1	250 500	10 to 10M 10 to 10M	240 180	720 500					
Tolerances available ±20% 10% 5%									
Н	IGH ST	ABILITY RE	SISTORS						
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Tolerances available ±5% 2% 1%									
	WIREW	OUND RESI	STORS						
5 & 10	_	5 to 100K	72	300					
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A.I.D & A.R.B - APPROVED

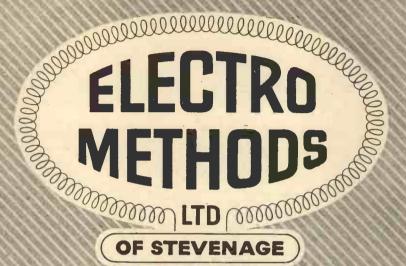
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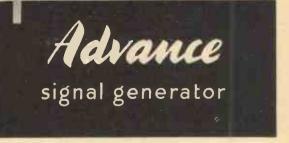
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Our chief Engineer, who is operating a Technical Guidance Service, is available daily including Saturdays from 10 a.m. to 6 p.m. or will deal with enquiries by return of post.

Our new illustrated Catalogue and supplement will be a great boon to those desiring quality equipment for modest expenditure. Send two 2½d, stamps for your copy now. It may well save you pounds! All our equipment is on demonstration at our showroom in conjunction with a variety of Pickups, Speakers, etc. If you can possibly call we shall be pleased to see and help you. H.P. facilities available.

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The New No. I "SYMPHONY" AMPLIFIER MARK III is a 3-channel 5-watt Gram/Radio Amplifier with astonishingly flexible tone control. You can lift the treble, the bass, or—and here is the unique feature—the middle frequencies to suit your own ear characteristics the middle frequencies to suit your own ear characteristics and the record or radio programme being heard. It is thus possible to arrange frequency-response of the amplifler to a curve equal and opposite to the resultant curve of the other items in the chain so what finally registers in the brain is as per original. This flexibility of control is even more important than the nominal linear response of the amplifier, as the pickup speaker, etc., are not linear. Independent Scratch-Cut is also fitted and special negative feedback circuit employed. The Amplifier can accommodate a wide variety of records from old 78s to new L.P.s, and there is full provision for Radio Tuner, Tape take-off and Playback. It is available to match 2/3 or 15 ohms speakers. Price 12 gns. (carriage 7/6). Fitted in portable Steel Cabinet, 2 gns. extra. in portable Steel Cabinet, 2 gns. extra.

output and triodes throughout. Woden mains and output transformers and choke, Output tapped 3, 7.5 and 15 ohms. Provision for Tuner and Tape.

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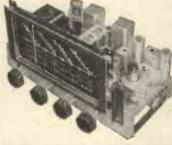
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N.R.S. EMPRESS FM/VHF TUNER/ADAPTOR Fine little job, w'll plug into any radio and add F.M. £13/15/-. Magic eye assembly £1 extra if required. Ditto mounted in beautiful dark walnut cabinet complete with magic eye 17 gns. Carriage 7/6.

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MIDAX/TREBAX CABINET. To match "Sherwood" (Viscount) £710/-.
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30in. high, consist of fully-cut £in. thick, heavy, inert, nonresonant, patent acoustic board, deflector plate, felt, all
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19in. wide x 13in. high x 13in.
deep and finished in polished
walnut, thus enabling a complete walnut, thus enabling a complete installation to be built up unit by unit in matching style cabinets and added to as required. Any of our Amplifiers, Tuners, Gram Units, Tape Decks and Speakers can be supplied in these cabinets for small extra cost. Examples on demonstration.

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2-speed, twin-track, microphone, radio and external amplifler in-puts. Facilities for playback through high quality internal elliptical speaker or through external high fidelity speaker or through external high fidelity amplifler. Automatic head de-magnetisation. Wide frequency range heads. Housed in hand-some polished walnut cabinet. Fantastic value for money at 49 gns. or 9 monthly payments of 6 gns. De luxe model with built-in revolution counter, 52 gns. or 9 monthly payments of £6/13/0. Plus carr. and pkg. £1. Full details in catalogue.

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Full particulars of this tube and other units specially designed for use in the higher tropospheric scatter and other microwave equipment.

frequency bands are available on application.

-				1					
	E.E.V. Type	Function	Centre Frequency (Mc/s)	Maximum Output	Noise Factor (dB)	Gain (dB)	Helix Volts	Collector	Focusing Field (Gauss)
	N.1001	Power	2000	16W	_	26	2600	40mA	600
W	N.1002	Low Noise	2000	ımW	10	24	550	200μΑ	200
	N.1004	Power	4000	4W	-	21	2600	20mA	450
	N.1005M	Low Noise	4000	ımW	II	22	360	200μΑ	350
	N.1013	Voltage Amplifier	2000	200mW	20	32	650	4mA	300
	N.1017M	Low Noise	1200	1mW	10	20	700	200μΑ	200
	N.1018M	Voltage Amplifier	4000	100mW	20	30	630	2mA	350

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50 micro-ohms to 10,000 megohms ·0002 picofarad



THE B.221 is a highly accurate transformer ratio arm bridge of very advanced design. It provides facilities for the two, three, or four-terminal measurement of impedance or transfer admittance over an extremely wide range at an operating frequency of 10,000 radians/sec. (1592 c/s).

Measurement is unaffected by the impedance of the test leads, which can therefore be of any length. Consequently the instrument is ideally suitable for the determination of temperature coefficient of components under test conditions or, in fact, any remote in-situ measurement.

A novel mechanism automatically displays the cyphers, decimals and units of measurement. This gives direct reading and avoids any confusion which might be caused by the large multiplying factors involved.

The basic range of the instrument covers im-

pedances from 10,000 megohms to 10 ohms and this is extended to 50 micro-ohms by the use of the Low Impedance Adaptor. Other adaptors have been designed for measurement of conductivity, dielectric constant and loss factor of solids and liquids.

SPECIFICATION

BRIDGE ONLY: Capacitance: 0.0002 pF to $10~\mu\text{F}$ in 7 ranges. Accuracy $\pm\frac{1}{4}\%$. Conductance: $0-\pm100$ mmho in 7 ranges. Inductance: 1 mH to infinity in 7 ranges. Measuring Frequency: 10,000 radians/sec. (1592 c/s.) Power Supply: 110/115 and 200/250~V 40/60 c/s. Dimensions: $17\text{m} \times 7\text{m} \times 11\frac{1}{4}\text{m}$ high. Weight: 25 lbs. approx.

WITH LOW IMPEDANCE ADAPTOR: Capacitance, $I_{\mu}F = 100,000 \mu F$ in 4 ranges, Resistance: $0-100 \Omega$ in 4 ranges. Discrimination on lowest range $50 \mu \Omega$. Inductance: 0-10 mH in 4 ranges. Discrimination on lowest range $5m \mu H$.

PRICES: Bridge, £175. Low Impedance Adaptor £15.

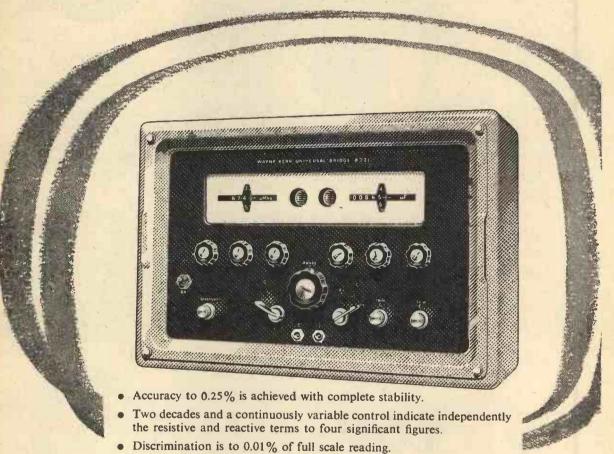


WAYNE KERR

THE WAYNE KERR LABORATORIES LIMITED,

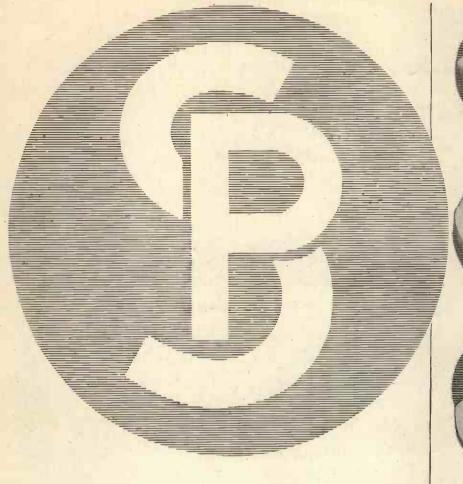
and Chemists

to 100,000 microfarads, 5 millimicrohenries to infinity



- Test leads of any length do not affect accuracy.
- Transformer ratio arm principle allows measurement in any quadrant of the complex plane.
- Adaptors for measurement of conductivity, dielectric constant and loss factor of solids and liquids.
- Lowest capacitance range is 0-1 pfd. with 0002 pfd. first scale reading.
- Low Impedance Adaptor extends resistance and inductance range to 50 micro-ohms and 5 millimicrohenries. This covers:
- Dry joints, earth-bonding efficiency, switch contact resistance, cable connectors, valve base connections, and general circuit strays.

ROEBUCK ROAD, CHESSINGTON, SURREY. TELEPHONE: LOWER HOOK 1131



for radio ceramics





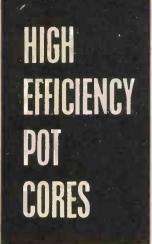






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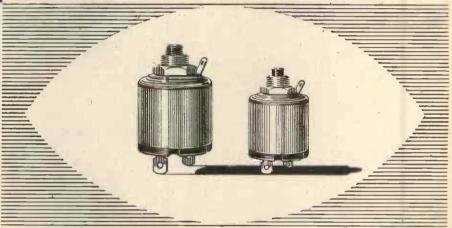


CAN NOW BE ADJUSTED.....

AFTER MOUNTING

OVER AN AVERAGE RANGE OF 10%

WITH A SETTING ACCURACY OF .5%



These new Mullard 14mm and 18mm pot cores are completely self contained, simple to mount and easily adjusted after mounting. Unique features include single hole fixing and two or four way terminal plates. Adjustments of inductance are made by means of a screw which varies the position of a magnetic shunt in the centre of the core; in many cases this eliminates the need for trimming capacitors. Designers will see from the brief characteristics listed here that Mullard adjustable pot cores are particularly suitable for use in high grade communications equipment, tuned circuits and filter networks. Those requiring further technical details are invited to write to the address below.

14mm Pot cores

18mm Pot cores

Four types available LA42—45

Air gaps From 0.3mm — 1.0mm

Frequency range 10 Kc/s — 100 Kc/s

Q values in the higher frequency range > 300

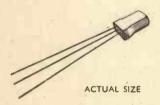
Mullard



'Ticonal' permanent magnets
Magnadur ceramic magnets
Ferroxcube magnetic cores

small junction switches Transistor... The ideal

XFT2



For Hearing Aids and other Audio Frequency applications

Incorporating a germanium junction element manufactured and supplied by



Encapsulated in a seamless metal can, its dimensions are only-4.75 × 3.1 × 7.5 mm.





M.30: 10-amp. Toggle Switch.

T.622: Toggle Switch, D.P.C.O., 3-amp., 250v.

S.510: D.P.D.T. Biased Push Switch.

T.600: 3-amp. Toggle Switch.

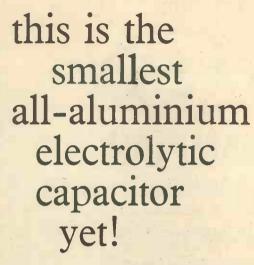
S.936: Push Switch, available normally On or Off.

Small Pointer Knob for 4in. spindles.

Write for Catalogue No. 129



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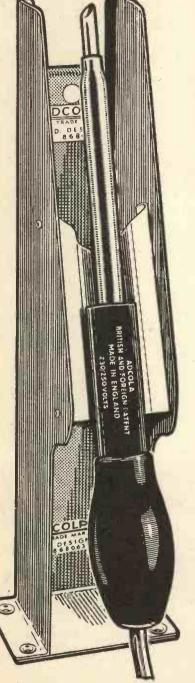
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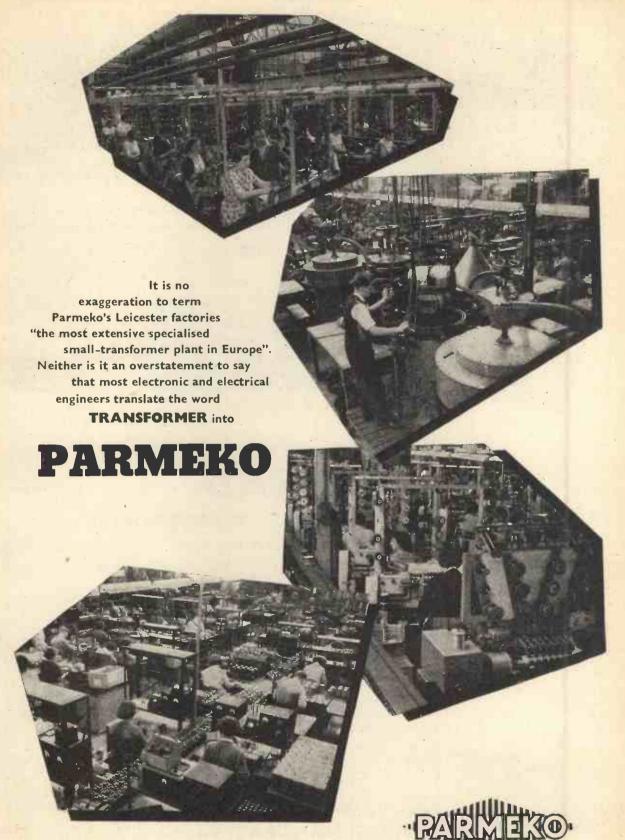
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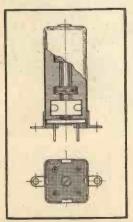




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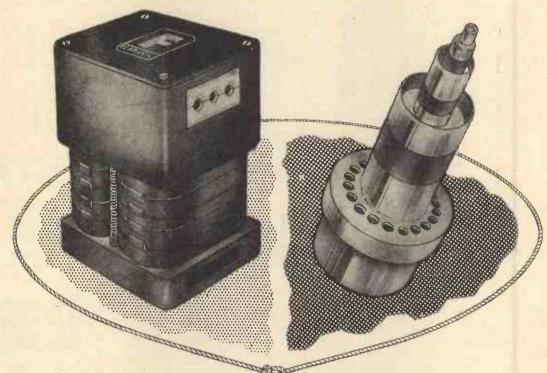
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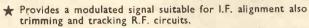
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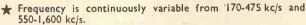
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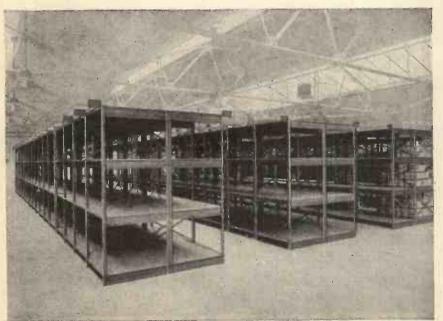
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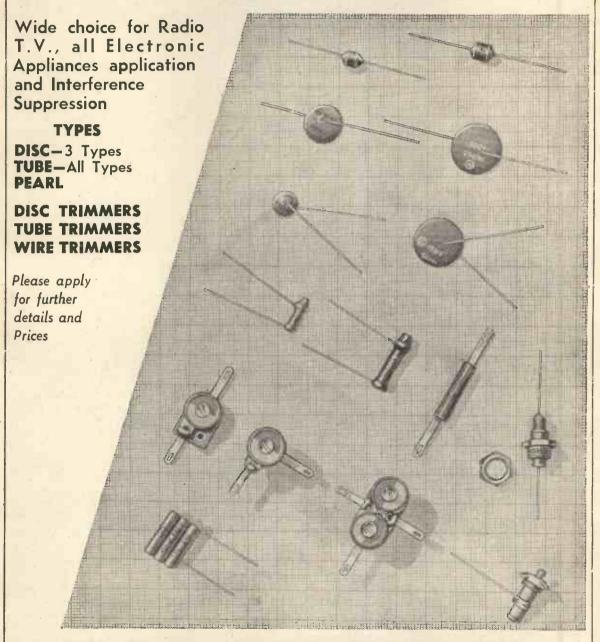
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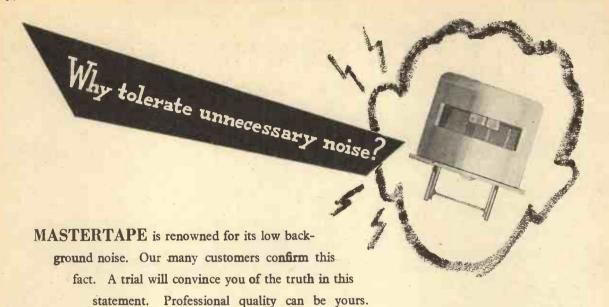
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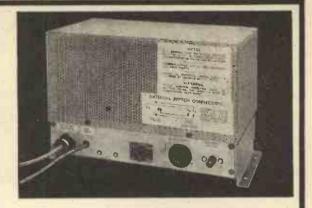
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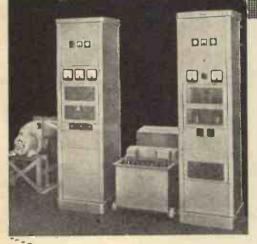
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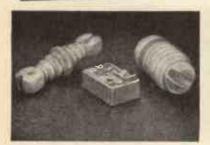


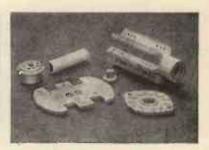
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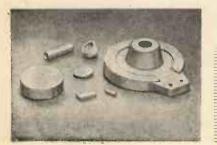
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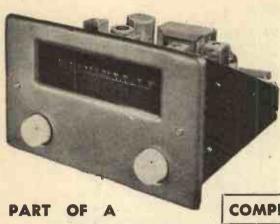
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Attenuation dB/100 ft. at 50 Mc/s.	3 · 0	3 · 4	2 - 3	2 · 6	1 . 5			
200 Mc/s.					3 · 3			
Copper Conductor	1/:022"	7/-0076"	1/:029"	7/:010*	1/:044"			
Diam in Inches:-								
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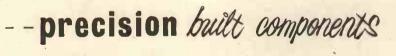
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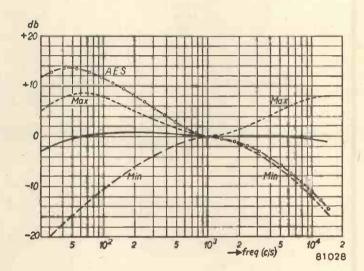




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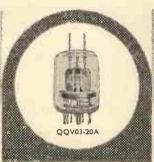
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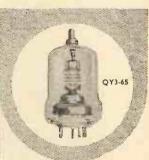
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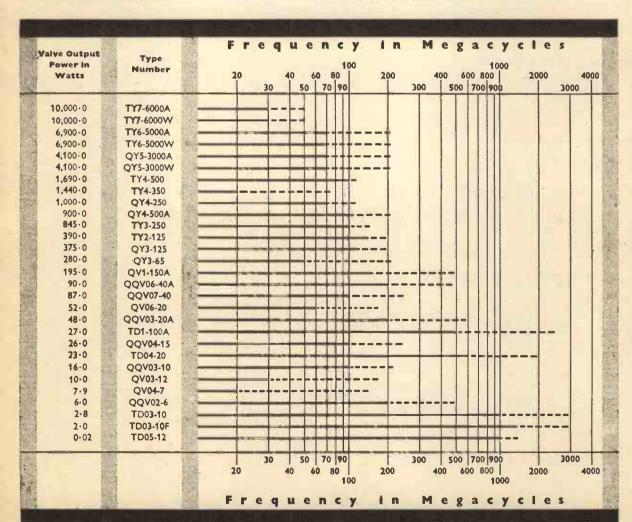
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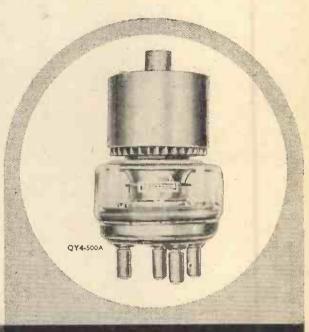
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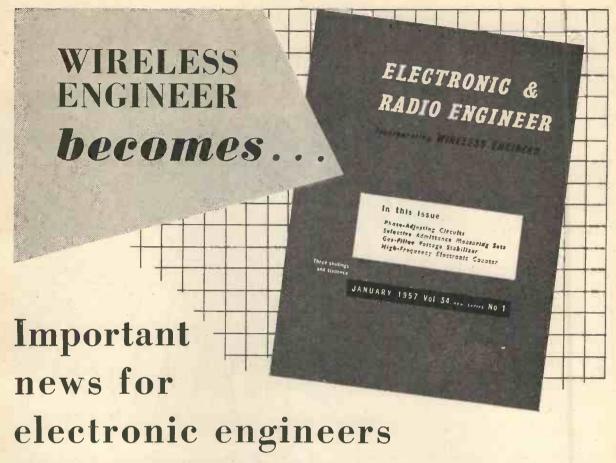
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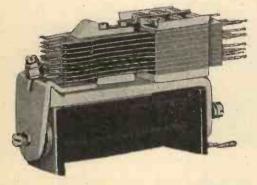
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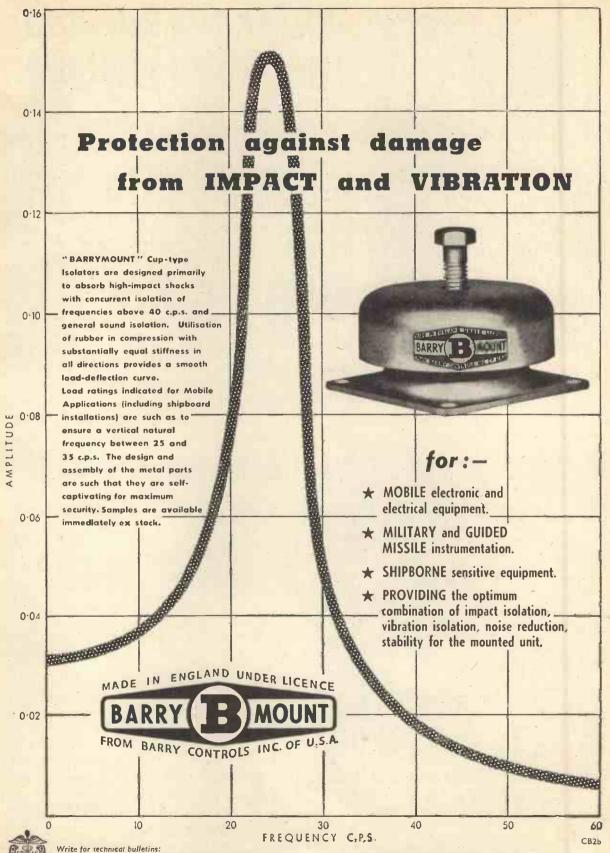
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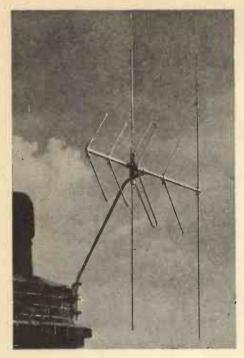
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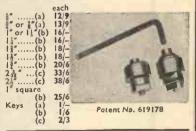
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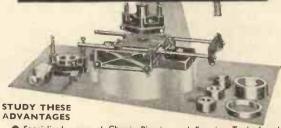
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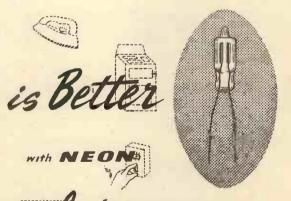
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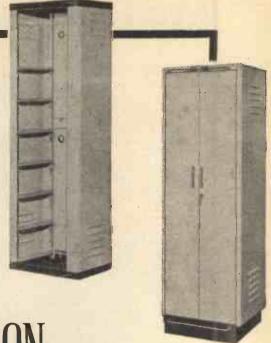
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Appendix

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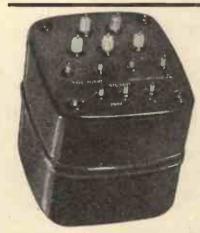
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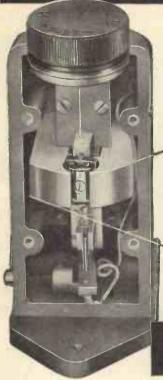
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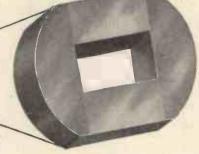
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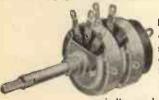
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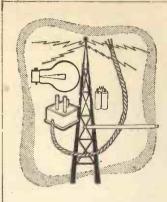
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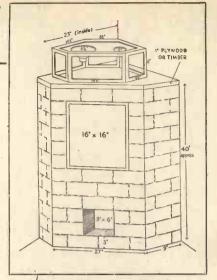
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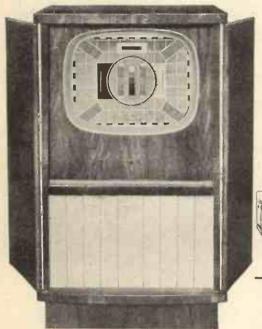
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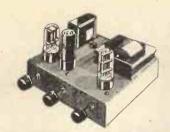
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Two colours, wine and grey, with cutout for speaker and under lid when closed 64n.

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As above but with cutouts for Speaker and Amplifier 79/6

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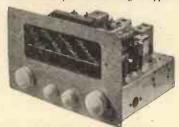
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Complete, ready to connect to any Type of Pick-up and Speaker (3 ohns) A.C. Mains 200/250 volts. Volume and tone control fitted with knobs. Overall size 7½m. Jong x 3½m, wide x 2½m high. £2.19.6

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PREMIER BUREAU DE LUXE

A superb cabinet in finely figured walnut veneer. Interior light sycamore, with rex-ine matching lining.

Overall dimensions:
33ln. high, 34ln. long,
174ln. deep. Uncut
control panel on
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size felt-lined storage cupboards in the lower part of the
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This converter hasbeendesigned to receive alternative programmes on Band 3 Channels 6 to 13, selection of Channel being made by

of Channel being made by rotation of Switch Knob. Mains supply 200/250 v. A.C. only. The high noise factor ensures good picture quality. A variable gain control makes it possible to balance a Band 1 and Band 3 signal inputs to the receiver. Any single channel Receiver of either T.B.F. or Superhet design may be fed from the Converter. Separate input sockets are provided for Band 1 and Band 3 inputs. Cash price £9/19/6. Credit Terms deposit £1/5/6 and 8 monthly payments of £1/5/6. IMPORTANT, please state your Band 1 station. Packing and postage 3/- extra.

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A really elegant cabinet in well figured walnut veneer. Suitable for



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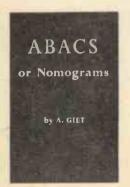
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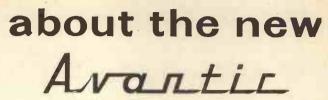
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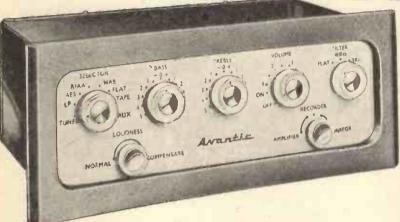
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The DL7-35 with wide range speaker system can be supplied in two superb cabinets finished in natural mahogany at £144.5.0. net. Provision is made for any of the following items which can be fitted as optional extras: 4-speed single optional extras: 4-speed single or automatic record player; Avantic vhf-fm or mw-am/ vhf-fm radio feeder unit; Avantic tape player. The Avantic loudspeaker system comprises a 12"

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Push-pull distributed load output stage producing an output of 27 watts at ± 0.1% total distortion.

Frequency response: ± 1 dB 1 c/s. to 100 Kc/s.

Damping factor: 50. Sensitivity: 255 mV. for 27 watts output. Hum & noise: —89 dB relative to 20 watts output.

Output impedances: 4Ω , 8Ω & 16Ω switch selected; automatic feedback adjustment. Built-in volume control and two audio input sockets.

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Output: 200 mV, at 0.1% and 2.0V, at 0.2% total distortion. Intermodulation distortion: power & pre-amplifier combined: 1% for 20 watts output.

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Managing Editor: HUGH S. POCOCK, M.I.E.E.

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FEBRUARY 1957

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A Transistor



Controlled Regulated Power Unit

The circuit diagram of the regulated power unit, shown below, demonstrates a novel application of transistors as control devices. They are especially sulted for this type of circuit because of the high voltage gains which can be obtained with them when they are operated with large collector loads from high voltages. In the power unit described, a Mullard OC73 is used to control the current flowing through an EL33 series valve. One advantage of using a transistor in this type of regulated power unit is that a separate negative h.t. supply is eliminated.

The power unit has an output impedance of less than three ohms, and will deliver 100mA at any voltage between 40 and 84 volts regardless of quite large changes in the mains input voltage. For instance; if the mains input voltage rises from 198V to 242V, the output voltage will increase by only 0.4V at full load. Reducing the load from 100mA to zero produces a rise in output voltage of only 0.3V at 220V mains input.

CIRCUIT OPERATION

By connecting the transistor as shown in the circuit diagram it compares the output voltage with the reference voltage, and any difference produces a large change in collector current due to its high effective mutual conductance.

Since a transistor can be operated at very low currents, very high voltage gain is obtained by connecting the collector to negative h.t. through a load of $500k\Omega$.

The collector is connected also to the control grid of the EL38 through a $100 \, \mathrm{k}\, \Omega$ grid stopper. Therefore the collector-to-base voltage of the transistor is equal to the grid-to-cathode bias of the valve and is practically independent of the output voltage setting.

The output voltage is approximately equal to the reference voltage and can only differ from it by the base to-emitter voltage of the transistor, which will be less than 0.1V. So the maximum output

which can be obtained is determined only by the reference voltage. Although an 85V reference level is used in the circuit described, the design is almost identical for any reference level. The minimum output voltage of this supply unit cannot be set below the bias voltage of the series valve and therefore it is limited to about 40V.

When the power unit is working, a drop in output voltage will cause the base of the OC73 to become negative with respect to the emitter. As a result, the collector current increases and the collector voltage becomes more positive, thus reducing the bias on the EL38 and compensating for the original change. Equilibrium will be reached when the output voltage reaches the same value as the emitter voltage.

Variations in the mains voltage have only a slight effect on the reference voltage, which is stabilised by the 85A2, and therefore the output voltage remains substantially constant.

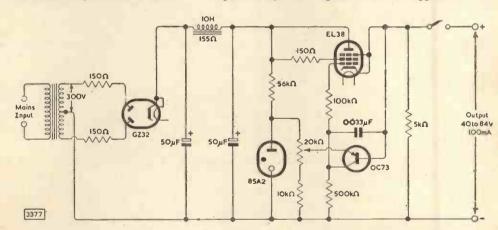
CIRCUIT DESIGN

The full-wave rectifier circuit is of conventional design, with choke smoothing to give a ripple-free output to the series valve, an EL38, and to the voltage reference tube, an 85A2. The 85A2 is operated at a low burning current to minimise the change in the reference voltage when the circuit is loaded and the poor regulation of the rectifying circuit causes the unstabilised voltage to fall.

An EL38 was chosen for the series valve because it has a comparatively short grid base. This is important since it governs the maximum collector-to-base voltage when no current is being taken from the supply. To prevent the collector voltage exceeding the $-30\mathrm{V}$ collector-to-base rating of the OC73, the output must be permanently loaded to about 10mA; $5\mathrm{k}\,\Omega$ forms the permanent bleeder resistor.

The stability of the output is limited by the stability of the reference source, and by using an independent reference voltage or running the 85A2 from a 150B2, the stability could be increased, and the output impedance reduced to at least half the value given.

Changes in temperature have negligible effect on the output voltage and impedance but an appreciable in-



crease in ambient temperature raises the minimum transistor leakage current and slightly increases the minimum output voltage to which the supply can be set. This effect limits the maximum value of collector resistance which can be used. For operation at higher temperatures a smaller collector load resistance should be chosen.

Instability may occur when the power supply is first switched on if it is connected directly to a high current load, so a switch has been included in the circuit and should not be closed until the reference tube has ignited. The 0.033 µF capacitor between the base and collector prevents high frequency oscillation which can occur under certain conditions.



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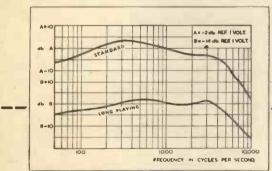




Acos G.P. 61

ceramic pick-up cartridges pass B.S.I. tropical test with flying colours

A batch of ACOS GP61 Ceramic Cartridges was recently tested by an independent user under the BSI Test No. BF 2011, Class 112, "Basic Climatic and Durability Test for radio and allied equipment." The result shown on the right speaks for itself. Further, the GP 61 has great mechanical robustness, smooth response, low harmonic and intermodulation distortion, high needle tip compliance, replaceable "X 500" styli and very good output.





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"BELLING-LEE" NOTES



Photo by courtesy of the Scotsman.

A recent photograph of the I.T.A. site at Blackhill showing the commencement of the main mast on the right, and in the centre the "Belling-Lee" "Skytower" acquired by the Authority and which will be used for the early test signals. This mast has now been completed and signals scheduled to commence early in March. On the left is Mr. A. P. Hale, Grad.I.E.E., and Mr. Boyd, both of Belling & Lee, whose riggers are seen at work. The sheet of water in the background is the Airdrie Coatbridge and District Reservoir.

In these uncertain times it is always dangerous to look into the future, but many Scottish readers will be glad to know that an alternative television programme is definitely on the way. The "Belling-Lee" "Skytower," recently purchased by the Independent Television Authority is seen being erected. It was completed before the photograph was printed. This aerial will be used for the pilot transmitter, scheduled to be on the air in March. The Scottish public will be given the opportunity of " seeing what they are buying," what kind of programmes they can expect. Scottish Television Ltd., the programme company, is organising a series of exhibitions throughout the service area commencing in February.

Radio dealers will be able to tell their friends all about them, and when one will be in any particular district. The dealer will also be able to give guidance on the type of aerial that will probably be required or what modifications will be needed to an existing B.B.C. aerial. Most "Wireless World" readers know that the reception of band III presents greater difficulties than the reception of band II. For example on band III the loss in any cable is about twice as much as it is on band I. It is important only to use solid cable on short runs. Cellular type is recom-

mended for general use, and semi-air-spaced for long runs or anywhere outside the transmitter service area. On a 50 foot run, the gain obtained by using air spaced feeder instead of solid is 1.7 dB which is approximately the difference between a 3 and a 6-element array. The difference in price between the aerials is 18/-, the difference in price between 50 foot of solid and semi-air-spaced feeder is £1/3/4. The superior aerial can help you to get rid of ghosts and interference, the better feeder can only help with gain. In difficult locations the better aerial should be specified and the better feeder.

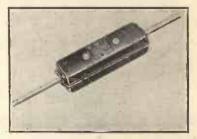
Radio Wave Propagation and the problems of Television Bands IV and V.

By R. L. SMITH-ROSE, C.B.E., D.Sc., F.C.G.I., M.I.E.E.

In a recent issue of the Journal of the Television Society a paper was published under the above heading. It includes a wealth of information of great value to all technical readers interested in television signal propagation. We are indebted to the author and the Society for permission to reprint this paper, a copy of which will gladly be sent on request.

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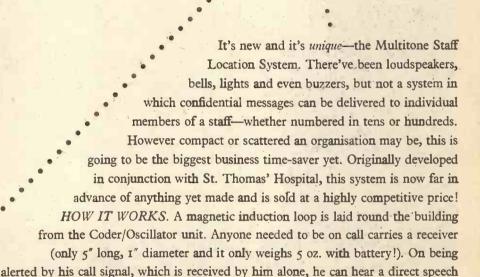
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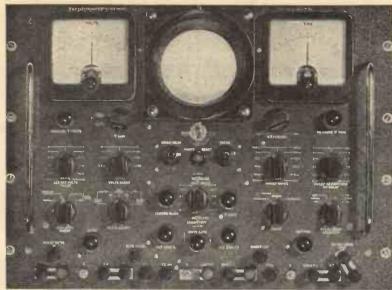
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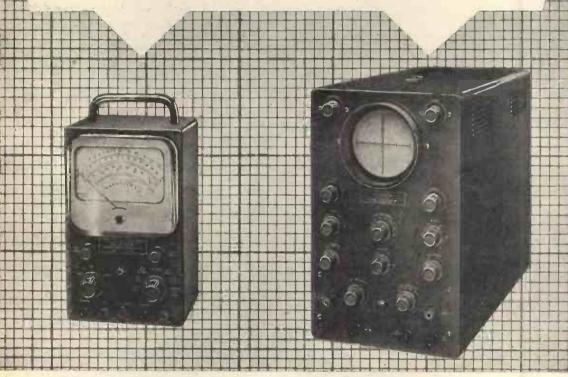
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MECHANICAL DATA

Maximum overall length	66.7	mm
Maximum seated height	60.3	mm
Maximum diameter	19.1	mm
Base	B7G	
Net weight	8.5	g

CHARACTERISTICS

Maximum striking voltage	90	V
Stabilising voltage	55	٧
D.C. operating current 2 to	30	mA
Maximum peak current		
(10 seconds max.)	75	⋅mA
Nominal regulation 2 to 30 mA	3	٧
Maximum regulation 2 to 30 mA	5	· V
Ambient temperature range		
-55 to +	90	°C

Type



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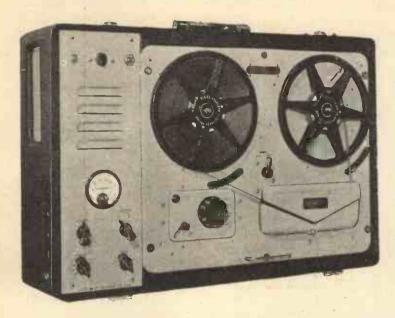
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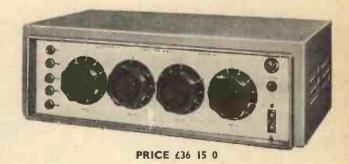
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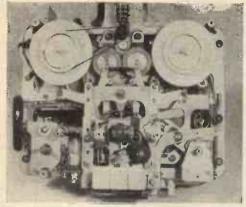


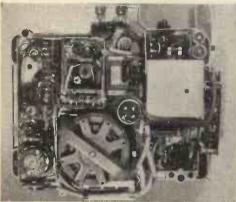
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Suitable for A.C. only, 105-115, 190-210, 210-230, 230-250 volts, 50 c/s.

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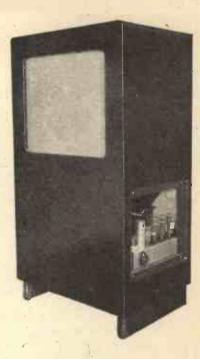
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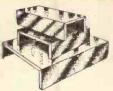
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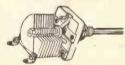
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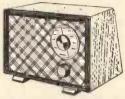


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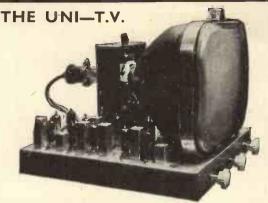
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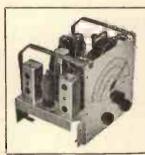


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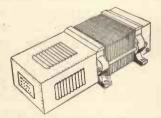
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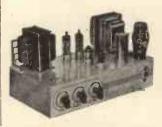
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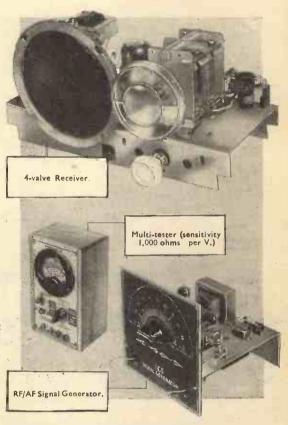
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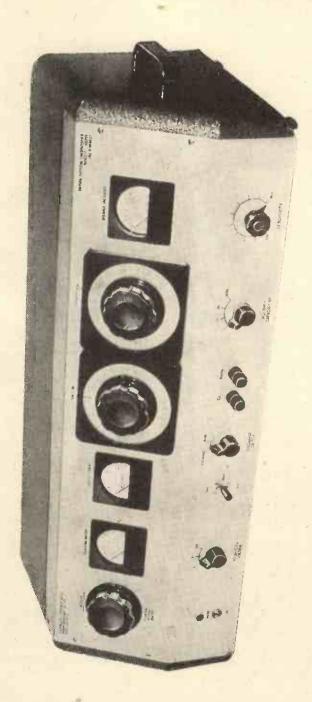
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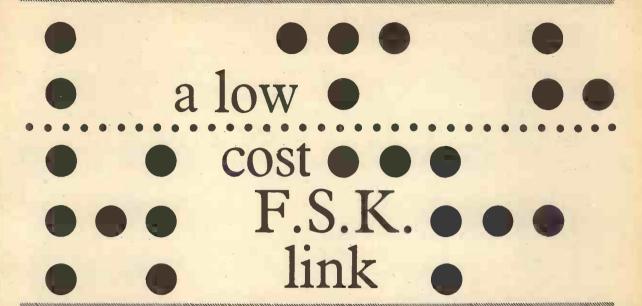
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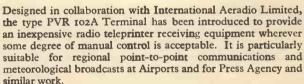


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Due to its low initial cost and its relatively simple installation, the terminal now offers radio teleprinter communications for use in circumstances where financial considerations have previously made them difficult to justify.



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Built-in metering arrangements provide for checking, setting-up and monitoring of all important circuit functions.

Direct operation of up to three teleprinters; no keying relays are used, thus ensuring maximum reliability and freedom from radio interference. Five separate channels, each pre-set and crystal controlled. Provision of crystal trimmers ensures that frequency is exact. Change of channel is effected by the operation of one knob.

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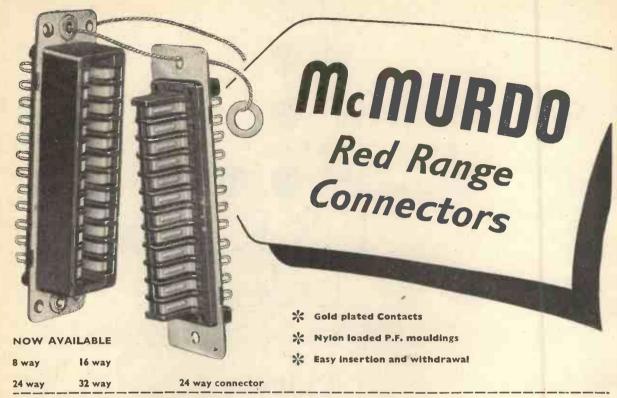
F.S.K. transmissions of differing shifts catered for by the use of plugin filters. Standard shift 200-850 c.p.s. by use of linear discriminator.

Rapid diversity switch action on small signal differential over wide range of inputs. Complete supression of weaker diversity of actions of the diversity of actions of the diversity of actions.

A brochure setting out extensive technical data concerning the PVR 102A Terminal is available on request. Please ask for Plessey Publication No. 789/1.

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Is the name of the new SONOMAG Pre-Amplifier recommended on page 238 of the November "Hi-Fi News" to those already owning Hi-fi equipment and wishing to add tape reproduction of the same quality.

This is the ONLY pre-amp, at present available designed specially for the new Collaro Transcriptor, and rigidly fixed as a

Demonstrations to all hi-fi enthusiasts of our pre-amp, used in conjunction with the Collaro Transcriptor Tape Unit, Collaro Transcription Motor, Leak Dynamic Pick-up and Diamond stylus, Leak Trough-line F.M. Tuner, Wharfedale Baffle 3-speaker system and Leak main ampliflers, will convince you of the fine standard of recording possible. Day, or evening (by appointment).

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Fitted into Fireside Console cabinet, oak, walnut or mahogany finish, 40 gns.

Your own Collaro Unit fitted, aligned, tested and guaranteed (at our factory only) for 16 gns.

Complete Tape Recorders, including Collaro Microphone and 200ft tape. Portable 48 gns. Console (with extra large speaker)

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T.O. CRYSTAL CARTRIDGES B.S.R. complete with two styli. LASKY'S PRICE 18/6, post free.

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TRUVOX "SENIOR" **SPEAKER** DRIVING UNIT (Pressure type) WELL BELOW HALF PRICE!

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59/6 Carr. 5/-.

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B.S.R. type TU.8, as illustrated.
3-spd. motor and pick-up with HGP.59
t.o. crystal complete with styli.

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Strongly made and finished with light grey leatherette with black/silver bands, grey leathcrette with black/silver bands, finger-shaped handle for easy carrying, strong hinges and clasps. Pocket in lid for tapes and spools, door for microphone, etc. Overall dim: 15½in. long, 13½in. wide, 9in. deep. Inside dim: 14½in. × 12½in. × 5in. A real general purpose job; will take almost any type of Deck including Collaro Tape Transcriptor, also almost any type of Record Changer including RC.80, etc.

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LARGE SELECTION OF CABI-NETS for TV and radiograms, carrying cases for record players, tape recorders, etc. Your enquiries invited.

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You can build it complete with valves and for only

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10 STAR FEATURES make this the finest portable radio ever offered to the home constructor. Peak value for money has been achieved without any sacrifice of quality or design.

SIMPLICITY OF CONSTRUCTION. The use of all the latest innovations gives maximum simplicity of construction combined with fine quality performance. In particular, the PRINTED CIRCUIT greatly simplifies construction and completely eliminates the possibility of wiring errors. The veriest novice can build this portable radio easily and quekly and the cost complete, less batteries, is only £7/7/-. Plus 3/6 postage.

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* PRINTED CIRCUIT 71in. × 21in.

★ 4-valve Superhet, med. and long waves.

Low consumption Double Battery Life. Valves.

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New style Contemporary Case. Lightweight and Handsome

Appearance. ★ Every Component available separately

YOU CAN BUILD A POWER SUPPLY UNIT FOR THIS PORTABLE.

For 200-250 v. A.C. mains. Made for the new low consumption valves. COMPLETE KIT containing printed circuit, brand new TCC condensers, latest STC metal rectifiers, all components, and full instructions, 45/-. Post 2/-. Also suitable for most other portables. Full details on request.

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A REALLY FIRST CLASS F.M. TUNER FOR HOME CONSTRUCTION ON A PRINTED CIRCUIT

Note these star features:—

HIGH SENSITIVITY.

ALL BRAND NEW T.C.C.
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AERIAL COIL AND R.F.
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\$ 5 VALVES AND 2 GERMANIUM DIODES.

By the use of a printed circuit the I.F. and R.F. amplifiers are extremely stable at maximum gain and results are consistent of the contract of

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R.F. Amplifier, Z719 or BF80.

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Write for full instructions data and illustrations, 2/6 post free.

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A super-sensitive Tuner for F.M. and medium waves. The complete parcel with power supplies. Post 3/6
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Ideal as radio receiver for inclusion
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A.C./D.C. 200/250 v. I.F. 465 kc/s.
A.V.C. 4 watts output, 3-station pre-set,
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Two controls, volume and station switch.
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Famous make. Covers Bands I and III.
Complete with valves EF80 and EC81.
Ceramic valve holders, finest quality
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LASKY'S PORTABLE TAPE RECORDER KIT

Comprises:

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★ Latest type Truvox Deck 23 gns.

* Tape Recorder

Amplifier 12 gns. 7×4 Elliptical Speaker,

★ 1,200ft. E.M.I. Tape, * 7in. Plastic Spool, 4/6

> LASKY'S PRICE COMPLETE KIT



36 GNS. Carr. & Pkg. 25/-.



LATEST BRENELL TAPE EQUIPMENT

Deck. 3-speed, 3 motors, record and play
back. 18 gns. Deck. 3-speed, 3 motors, record and play back. 18 gns. Amplifier Mk. II. 5 watts, for use with 3 ohms speaker. Magic eye. 13½ gns. Carrying Case 24/18/-. Complete equipment less mike, with tape, in carrying case ready for use.

Full details on request.

48 GNS.

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Contains everything to build a 4-valve 3-wave superhet for 200/250 A.O. mains. Uses 6KS, 6K7, 6C7, 6V6 valves. Attractive wood cabinet, walnut veneer, or plastic cabinet as illustrated. Size 12×6j x5jin. deep. OAN BE BUILT FOR £7.19.6 Carr. and packing 2/8.

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Contains everything to build a T.R.F. S-walve set for 200/250 A.C. mains, medium and long wave. Uses 6KTG, 6J7, 6V6 and metal rectifiers. Neat plastic cabinet, wainut or lovey finish, or wood cabinet. Size 13x 6jx6jin. deep.
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All components available separately.

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S.W.G. undrilled. 4 sides, 18 S.W.G. unarilied, 4 sides, reinforced corners. Depth 2\(\frac{1}{2}\)in. 6×4 4/-; 12×8 7/-; 16×10 8/3; 8×6 5/-; 14×9 7/6; 12×3 4/9; 10×7 6/-; 16×9 8/-; 12×6 6/6. Post 1/- per chassis extra.

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Anti-clockwise, shaded pole. Col-laro 25/-. Garrard 26/6. B.T.H. 29/6. Post extra.

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"Ferrovoice," on plastic spool, 1,200ft. List, 32/6. LASKY'S PRICE 25/-Post 1/-.

Spare 7in. spools, 3/6.

SPECIAL OFFER! Magnetic Recording Tape, kraft base. On Cyldon metal spools: 1,200ft. 11/6; 600ft. 7/6 On plastic spools: 1,200ft. 12/6; 600ft. 9/6.

PURETONE Tape plastic spool, 1,200ft., 14/11. Post 1/- extra.

All makes of Tape including thin long-playing. Also spools.

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Med. and long waves, wound, ready for use. Each 6/9, post 1/-

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5in. long, in. diam., with full instructions for making a Ferrite rod aerial, 2/6, post 1/-

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All 200-250 v. 50 e.p.s. primary, finest quality, fully guaranteed.

MBA/3. 350-0-350 v. 80 mA. 6.3 v. 4 a., 5 v. 2 a. Both filaments tapped at 4 volts.

MBA/7. 250-0-250 v. 80 mA. 6.3 v. 3 a. 5 v. 2 a. Both filaments tapped at 4 volts. 19/6.

AT/3. Auto trans. 0-10-120, 200-230-240 v. 100 watts. 19/6. MT/340. Tapped input 200/250 v. 300-0300, 100 mA., 5 v. 3 amp., 6.3 v. 1.5 amp. 16/6.

MT/341. Tapped input 250-0-250, 120 mA., 6.3 v. 5 amps., fully shrouded. 27/6.

6-VALVE RADIOGRAM CHASSIS COMPLETE WITH VALVES

Famous Manufacturer's Surplus 6-valve 3-wave Superhet. 13-50 m. short. 200-550 m. medium. 1,000-2,000 m. long. 550 m. medium. Brand new Muliard valves: ECH42, EF41, L63, EB41, 6V6, g.t. EZ40 and finest quality components. Gram., switch, 465 Kc/s. I.F. tone control, three-colour dial-Overall size 13 x 5in. height 12 in. Aperture required for dial and controls 11 x 3 in. Complete with valves, output trans., | nobs. etc.

LASKY'S PRICE £10.19.6

Carriage and Packing 7/6 extra

5-VALVE RADIOGRAM CHASSIS complete with valves, £9/19/6. Carr. & Pkg. 7/6.

MOVING COIL P.M. SPEAKERS

2½in. 3in. and 3½in., 19/6. 5in. | 6½in. | 8in. | 10in. | 12in. | 16/6 | 17/6 | 25/- | 32/6 | 29/6 6in. with transformer .. 21/-7×4in. Elliptical..... 19/6 10 × 6in. Elliptical 32/6

GOODMANS 12in. AUDIOM 50 P.M. SPEAKERS

10 watts. Limited number only. Listed at £6/15/-. LASKY'S PRICE 97/6 Post free

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For high fidelity sound repro-duction. Easy to fit to any radio, TV receiver or amplifier. Full data and circuit diagram supplied with each.

LSH75. For outputs up to watts, 8/-. LSH518. For outputs of 10-12 watts, 12/6.

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FEW ONLY Type C17FM, rectangular, aluminised, 6.3 heater. Brand new and

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OC70 | OC71 | OC72 21/- | 24/- | 30/-

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TS1, 18/-. TS2, 21/-. TS3, 24/-. TP1 or TP2, 40/-.

TRANSCRIPTION TURNTABLES

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RIISS RECEIVERS Available on Easy Terms. In original wood transit cases.

Used, Grade 1, £9/19/6. Used, Grade 2, £7/19/6. Carr. 12/6.

Power Pack and Output Stage with 6 in. Speaker, 5 gns.

FAMOUS AMPLIFIERS BUILT ON T.C.C. PRINTED CIRCUITS

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All specified components are used and you have your choice of transformers and chokes by Partridge, Haddon, W/B, Ellison or Gilson. Demonstrations given at any time.



MULLARD 510. Fully assembled MULLARD 510. Fully assembled complete with valves ready for use. Price, according to make of transformers, from ... 15 Gns. COMPLETE KIT of parts and Printed Circuit for building the Mullard 510, from ... 12 Gns. Details on request. Book, 3/6 post free. Printed Circuit separately, 22/6.

OSRAM 912, fully assembled complete with valves ready for use. Price, according to make of 18 Gns. COMPLETE KIT of parts and Printed Circuit for building the 15 Gns. Details on request. Book, 4/- post free. Printed Circuit separately, 50/-.

All components for either above Amplifiers supplied separately. Price List on request.

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All 200-250 v. finest quality,	
6.3 v. 1.5 amp 6.3 v. 3 amp	 5/11
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WIDE ANGLE 38 mm.	
Line E.H.T. trans., ferrox-cube	25/-
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Line Output Transformers. No	
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frame	12/6
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Here are a vew examples of brand

new s	urpiu	s and	ımpoı	ted va	ives:
EB91	7/6	EV41	10/6	EY51	12/6
EB41	7/6	EF80	10/6	EC84	11/6
EABC8		EF85	10/6	EY86	14/6
EAF42	10/-	EF86	12/6	EZ40	8/6
EBC41	10/-	EF89	10/-	EZ80	8/6
ECC85	10/-	6K8	10/6	PCF82	12/6
ECC84	15/-	6₹6	8/6	PCC84	12/6
ECC83	9/-	6K7	5/8	PL81	13/6
ECC82	9/-	6Q7	10/6	PL82	10/6
ECC81	9/-	6R7	6/6	PL83	11/6
12AT7	8/6	5 Z 4	9/6	PY80	10/6
12AU7	8/6	DAF9	6 10/-	6AT6	7/6
12AX7	9/6	DL96	10/-	6AT7	7/6
ECF82	15/-	DK96	10/-	185	7/6
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ECL80	11/6	DM70	9/-	IR5	7/6
Also fr	If stock	e of B	V.A V	alves ar	dCR

Tubes at the new lower list prices. WRITE FOR COMPLETE LIST

* LASKY'S ANNOUNCE PORTABLE GRAM AMPLIFIER

Of very small dimensions and suitable for any type of mains operated portable Record Player. All brand new components, latest circuit technique. The price of the Kit, complete with valves, rectifier and 6 x 4in, elliptical speaker, will be under £4.

Full details on request.

DATA BOOK and shopping list, 1/6 post free.

LASKY'S RADIO

LASKY'S (HARROW RD.) LTD H.P. TERMS AVAILABLE Write stating requirements.

BAND III AERIALS. All types, outdoor or indoor, also Diplexes, Crossover Co-axial Plugs, Socket and Cable.

TWO ADDRESSES FOR PERSONAL CALLERS
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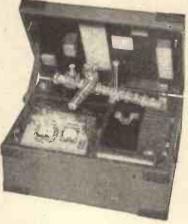
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There is always a fine selection of equipment at

3 cm. TEST SET

TYPE 263

Containing transmission type w/meter complete with detector unit 9280-9480 Mc/s, attenuator unit, 2 coaxial to waveguide feeders, impedance matching unit, medium power dummy load, standing wave indicator with lock using CV.263 indicator valve, metered indicator unit, various connectors. Suitable for testing medium and low power radar installations. Price £20 carriage paid.



WAVEMETER TYPE W.1310

Coverage 155 to 230 Mc/s. continuous, complete with Test Prod. Input 230 v. 50 cycles. New condition, £3/10/-, plus 7/6 carr.

BLOCK CONDENSERS

8 mfd. 600 v. W., 5/6 each, post paid. 4 mfd. 400 v. W., 4/- each, post paid.



AN/APN.1 TRANSDUCER

This Unit consists of Magnet, and Coil which is attached to an aluminium diaphragm sus. pended freely and perforated to prevent air Mounted on a Ceramic cover which sits over the diaphragm is a form of 2-Gang capacitor which has a swing from 10-50 pF.

The above unit is used as part of Wobbulator described on page 252 of the June "Wireless World." PRICE 7/6 p.p.

APQ.2. RADAR/JAMMING UNIT

Freq. 450-710 Mc/s. Containing 931a Photo Multiplier Cell complete with resistance network and light proof box. Wide band amplifier 2 6AC7, 1 6AG7, 2 388a. This unit is similar to the A.P.Q9 Jamming Unit. Brand new £5 plus 10/- carriage.

MORSE SIGNALLING LAMPS

(Aldis type) 5in. dia. with sighting arrangement, 2 handles, keying cwitch, and 2 yards cable. In wood carrying case, 10/- plus 3/- p.p.

MINIATURE I.F. STRIPS

Size 103 × 21 × 3in. Frequency 9.72 Mc/s. 2 EF.92s and 1 EF.91 I.F. amps. EB.91. DET/ AGC. EF.91 AGC. Amp. and EF.91 Limiter. Circuit sup-

Price 8/- less valves. Post paid.



TRANSMITTER Type T1131-L

Frequency 100 to 156 Mc/s. Output 50 W. Crystal controlled. 200-240 v., 50 c.p.s. Power supply. Housed in 6ft. standard on 19in. rack. In new condition complete with valves.

Send for full details.

EDDYSTONE 358X

COMMUNICATION RECEIVERS (B34)

Range 40 kc/s. to 31 Mc/s. covered with 10 plug-in coils; only 4 coils available covering 600-1,250 kc/s., 1,250-2,100 kc/s., 2,100-4,500 kc/s., 9,000-22,000 kc/s. Selectivity: 2 kc/s at 2.5 db down; 5 kc/s at 35 db down; 150 c/s. at 5 db down with crystal. Supply required: 6 v. 1.4 a; 175/200 v. 65 m/a. CIRCUIT: variable mu pentode H.F. amplifier, triode-hexode frequency changer, two I.F. amplifiers (450 kc/s.), crystal filter, A.V.C./ detector/A.F. amplifier, output stage, B.F.O. valve check meter. £8.10.0 With power supply, plus £1 packing and carriage.

POWER UNIT Type 173

24 volt D.C. input, 120 v., 60 · m/a, output. Containing Vibrator Transformer, 12 volt Vibrator, two 120 volt Selenium Rectifiers. Chokes and Condensers. Size 10 in. x 6in. × 3in. Price 12/6 post paid.



BENDIX TRANSMITTERS

TYPE T.A. 12B Master oscillator type transmitter. 4 channel 40 W. operation provide telephone, CW or MCW in frequency ranges of 300-600 kc/s, 3-4.8 Mc/s., 4-6.4 Mc/s., 4.37-7 Mc/s. Each of the 4 channels has its own oscillator and uses a 12.5K.7. The 1PA stage consists of an 807, while the PA is two 807s in parallel. Size 10½in. × 6½in. × 15½in. Price £315/s. plus 10/s.care. Price £3/15/-, plus 10/- carr.

INVERTERS

Miniature 3-phase (ex-compass unit) 24 v. input with 17 v. 3-phase, 400 c/s. output. These have been used by model makers as motors and are known as the "5/- Motor." Will run quite successfully on 12 volts. 5/- plus 2/- p.p.

I.F. AMPLIFIER UNIT

460 kc/s, with IT4. Brand new and boxed. Fully screened in plug-in box. Size 21in. × 1in. × 41in. Price, with circuit, 10/- each, plus 1/- p.p.



500 ohm, 4 figure no reset; size 5 x 11/2 x 1in. 5/- each, p.p.



BUZZERS

6 volt A.C., with tone adjuster, size 2 × 1 × 2½in., 4/- p.p.

BOOST GAUGES

2in, dia.; suitable after minor adjustment as car induction manifold meter. 2/6 p.p.

INDICATOR LAMPS

American panel type complete with 6 v. bulbs in set of 4, 3 green jewels and 1 red jewel, 10/6 post paid.

R.F. UNITS

R.F.24 20-30 Mc/s. Switched Tuning. Valved 9/6 each.

R.F.25 40-50 Mc/s. Switched Tuning. Valved, 9/6 each.

R.F.26 50-65 Mc/s Variable tuning, valved. Damaged dials 20/- each: perfec: dials 25/- each.

Packing and postage 3/- each type



All these fine offers are on display at -

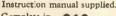


PROOPS BROS. LTD. -

The Walk-around Shop

VALVE TESTERS MODEL 314

This model is of American manufacture and versatile, free-point return valve tester. Its design is such tester. Its design is such that it enables the user to type valves, test any type valves, regardless of its filament regardless of its filament voltage or base wiring. Flexibility is attained by using individual lever switches for each valve element. Complete coverage of American Series including Acorns.



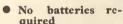
Complete in Carrying case £10 Plus 10/- carriage.

RIII5 RECEIVERS

Air Tested, in good secondhand condition. Price £6/5/-, plus 10/-packing and carriage.

A room - to - room telephone . . .

Ideal for two-way conversation, house-to-garage or internal communication.



quired No soldering required

Just connect it up and it works

The sets consist of 2 high-quality microphone/receivers (new and boxed) and 15 yards of twin wire.

COMPLETE FOR 8/6

plus 1/6 postage



MAINS POWER UNIT Type 234

(For use with Receiver T1392) Double Smoothed 200-250 v. 50c. input. 240 V. 100 mA. 6.3 at 6 amps. with Volt Meter reading input and output voltages. Size: 19in. × 10in. × 6½in. Standard Rack Mounting. Price £4/10/each, plus 7/6 carriage.



RECTIFIERS

Chassis cooled, brand new, 125 volt, 80 mA., 4/9 p.p.; 250 volts; 50 mA. 8/3 p.p.

BATTERY CHARGING LEADS

2 yds. of cab tyre twin cable, and 2 large crocodile clips; new and boxed. 3/- p.p.

HEATER TRANSFORMERS

6.3 volt, 11 amp. Brand new, 6/6 each plus 1/- p.p.

CR.300/1 RECEIVER & POWER SUPPLIES

Available for callers only.

DESYNN TYPE Antenna or Beam position indication

This comprises a Transmitter unit and Indicator which will operate on 12 or 24 volts D.C. and will indicate with instantaneous and smooth pointer movement. The Transmitter is a specially designed potentiometer and will operate the Receiver on a simple three-wire system and the receiver in this instance is calibrated in Gallons but dial could be easily altered to indicate a 360 Deg. sweep. Transmitter and Receiver with full instructions. Price 12/6 post paid.

THROAT MICROPHONES

Type T30. U.S. Manufacture

Complete with eleastic strap. Lead terminating at 2-pin plug PL.291, and socket JJ-048. New and boxed, 3/- each, post paid.

RADIO ALTITUDE METERS

5 mA. Basic movement, 3in. dia. circ. scale with approx. 300 deg. sweep. 6/- each, plus 2/- p.p.

TWIN COUNTERS

(Gallons gone) 24V, reading 4 figures and reset contained in housing, size 4in. dia. by 5in. long, 15/- plus 2/- p.p.

NICKEL IRON CELLS

1.2 volt, size 3\frac{1}{2}in. \times 2\frac{3}{2}in. \times 1in., unfilled 5/- each, plus 1/- p.p.

OCTAL PLUGS

(Bulgin bakelite type), 2/6 each p.p.

RECEIVER UNIT Ex 1143A

10.72 Mc/s. I.F.s. Frequency 100-120 Mc/s, suitable for conversion to 2 metres and Wrotham.

Owing to a large purchase we can offer these units fully valved with circuit diagram at 25/- each plus 3/- post/packing. Valve line-up: (4) EF50, (1) EL32, (2) EF39, (1) EBC33, (1) EA50.



2in. MAGSLIPS

50 v. 50 cycle transmitter and receiver units. Accurate to 1/10th deg. Guaranteed good working order, 35/- a pair, plus 3/- p.p.

STUD STITCHES

20 segment 5/16in. studs, base 5in. square with handle and housing. New and boxed, 5/- each, plus 1/6 p.p.

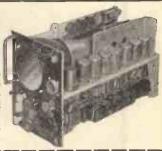
SPECIAL OFFER

MALLORY VIBRATOR PACKS

12 volt, 150 volt 40 mA. B: 5\frac{1}{2}in. \times 3in., 12/6 each p.p. Brand new and boxed, size 51in. X

TYPE 62A INDICATORS

Ideal for conversion to oscilloscopes, T.V. units, etc. Containing VCR.97, 12 VR.91 (EF.50), 2 VR.54 (EB.34), 3 VR.92 (EA.50), 4 CV.118 (SP.61). Slow-motion dial, 13 Pots and scores of useful components. Size 8½ × 11½ × 18in. In wooden packing case, £3, carriage 7/6.



NOTE: Carriage prices quoted apply only to England and Wales.

PROOPS

NOTE: Orders and Enquiries to Dept. "W". Shop hours 9 a.m. to 6 p.m.—Thurs.: 9 a.m. to 1 p.m.

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Stern NEW!! fidelity

TAPE RECORDER WITH EVERYTHING EXCEPT A HIGH PRICE

BEFORE CHOOSING YOUR RECORDER YOU MUST HEAR THIS NEW "fidelity" MODEL . . . IT HAS The BRENEIL 3-speed Tape Deck and a "fidelity" Tape Amplifier, based on a new design by the MULLARD TECHNICIANS and which we consider to be one of the best now available . . . Truly HIGH FIDELITY RECORDINGS are obtainable.

PRICE of COMPLETE RECORDER INCLUDING MOVING COIL MIKE and 1,200ft. **240'U**REEL of TAPE.
(Plus £1 carriage and insurance of which 10|- refunded on return of packing case.)

CREDIT TERMS. Deposit £11/10/- and 9 monthly payments of £4/4/4. HIRE PURCHASE. Deposit £23 and 12 monthly payments of £2/2/8.

-HOME CONSTRUCTORS-YOU CAN BUILD THE COMPLETE £42 RECORDER FOR

(Plus £1 carriage and insurance of which 10/- refunded on return of packing case.) CREDIT SALE TERMS. Deposit £10/12/6 and 9 monthly payments of £3/17/11. HIRE PURCHASE TERMS. Deposit £21/5/- and 12 monthly payments of

The BRENELL TAPE DECK and the "fidelity" TAPE AMPLIFIER are supplied tested and ready for use and the actual assembly of the recorder is extremely simple involving only a few connections for which a step-by-step chart is supplied.



- * High Quality Output Transformer by Gilson
- ★ 3 Speeds, 3½, 7½ and 15in., TWIN TRACK
- * Position provided for use as straight amplifier
- + Efficient Tone Control arrangement
- * High-grade Components throughout
- + Two position equaliser for 3% and 7% in.
- Monitor and Extension Speaker Sockets are provided
- Beautiful styling and cabinet

IF YOU HAVE YOUR OWN CABINET WE WILL SUPPLY . .

THE BRENELL TAPE DECK, the "fidelity" TAPE AMPLIFIER, MATCHED P.M. SPEAKER and 1,200ft. REEL PLASTIC TAPE.

Deposit £9 and 9 monthly payments of £3/6/-. H Deposit £18 and 12 monthly payments of £1/13/5. HIRE-PURCHASE TERMS. CREDIT SALE TERMS.

ALK £36'0'0

(Plus £1 carriage and insurance of which 10/- refunded on return of packing case.)

TAPE RECORDER EQUIPMENT-IN STOCK . . .

- BRENELL 3-SPEED TAPE DECK
- £18/18/-. The TRUVOX MK. III and MK. IV DECKS

- (fcon) £24/3/-.

 (c) The COLLARO TRANSCRIPTOR "TAPE DECK" £20 (see note below).

 (d) The LUSTRAPHONE MOVING COIL MIKE, HIGH IMPEDANCE, £3/7/16.

 (c) The ACOS GRYSTAL MIKE, HIGH (M-PEDANCE, £2/10/-.
- The 'fidelity' TAPE AMPLIFIER (as used in shore Tape Residen, fucluding matched P.M. Speaker, £16/16/-. (b)
- The ARMSTRONG P.A.B.O. TAPE PRE-AMPLIFIER, £12/10/- (suitable for nearly all tape decks).

STOCK.
NOTE. We recommend that a CORRECTLY MATCHED PRE-AMPLIFIER is purchased with the COLLARO Transcriptor. We can supply it with the pre-amplifier actually assembled on Tape Deck. £37/10/-.
HIRE PURCHASE and CREDIT TERMS are available on all above equipment.

STERN'S "fidelity" F.M. TUNING
UNIT A 5 Valve
Tuner incorporating
the latest Mullard Permeability Tuned Unit.
Price assembled less
Power Supply:

£14/10/0

(Plus 176 earr, and ins.)

TERMS: (a) H.P. Deposit £7/5/and 3 monthly payments of 18/4:
(b) Credit Deposit £7/18/6 and 9 monthly payments of £1/6/7.

Provides "Hi-Fi" "eproduction with any make of Amplifier and many Radio Receivers. It incorporates: 5 The latest Valve Line-up—ECOS5, 2 type EFS5, EF91 and EMS0.

A "Magic Eye" Indicator. 6 Power consumption is 1.7 amps. at 6.3 volts and 25 m/a. at 250 volts.

STERN'S "fidelity" COMBINED A.M. and F.M. TUNING UNIT

This is IDENTICAL to the Stern's F.M. Tuner illustrated above, but in addition incorporates the MEDIUM WAVE-BAND and thereby also provides a selection of foreign

stations.
Price £18/18/0 (Plus 7/6 carr.

TERMS:—(a) H.P. Deposit 29/8/- and 10 monthly payments of 21/1/-; (b) Oredit Deposit 24/15/- and 9 monthly payments of £1/14/7. Send S.A.E. if further data required.

The TRUVOX MODEL "C" TAPE AMPLIFIER, £17/17/-.

All types of PLASTIC TAPE by E.M.I., SCOTCH BOY and GRUNDIG, Standard and Long play ARE IN



Provides a maximum output of renowned Leak reception for precision engineering, fine appearance and fastidious wiring.

PRICE \$28/7/0\$

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TERMS: Credit: Deposit 27/7/- and 9 monthly payments of \$22/11/4, H.P. Deposit \$21/4/4- and 12 monthly payments of \$21/6/-. Send S.A.E. for fully illustrated leafet.

"COMPACT 5 " STERN'S **AMPLIFIERS**

EXPRESSLY DEVELOPED FOR VERY HIGH

REPRODUC-TION OF GRAM. RECORDS AND PARTICU-LARLY SUITABLE FOR HIGH QUALITY REPRODUCTION OF THE F.M. TRANSMISSIONS:

A 2-stage high sensitivity amplifier having SEPAR-ATE BASS and TREBLE CONTROLS and designed to give up to approx. 5 watts with very pleasing quality. PRICE £5/15/-.

The "Compact 5-2"

The "Compact 5-3"
A 3-stage version of the "5-2" model but in this case having an additional stage and incorporating negative feedback. PRICE FRISE. negative 1

The Amplifiers are compact and very attractively designed having a "Hammered/Gold" finish with a fully engraved front panel by which the entire Amplifier is conveniently mounted into a Cabinet, occupying no more space than a conventional Tone Control Unit. Send S.A.E. for illustrated Leafiet.

POWER SUPPLY. Is obtainable from a small separate Unit which apart from supplying power to either Amplifier, also has additional supply available for a Radio Tuning Unit. PRICE (additional to above). £2/10/-.

Open Monday to Friday 9 a.m.—6 p.m. Saturday 9 a.m.—1 p.m.

LIMITED

AMPLIFIERS PRE-AMPLIFIERS

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HIGH FIDELITY FOR THE ... HOME CONSTRUCTOR

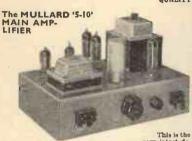
TUNING UNITS

COMPLETE KITS OF PARTS FOR THE "Hi-Fi" ENTHUSIAST-

FIER TONE CONTROL UNIT

"A design for the music lover"

QUALITY OF THIS NATURE HAS NEVER BEFORE BEEN OFFERED AT SUCH LOW COST. ISTERN'S "fidelity" PRE-AMPLI-



This is the very latest description of the very latest descrip

THE full SPECIFICATION and BUILDING INSTRUCTIONS for these three Units are available for 1/6 each. THEY include COMPONENT PRICE LISTS and simple "wire-to-wire" PRACTICAL D'AGRAMS.

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A COMPLETE KIT OF PARTS STERN'S "HIGH QUALITY" 8-10 WATT AMPLIFIER

Has power sup-ply available for Radio Tun-ing Unit. Price of COM-PLETE KIT OF PARTS

(plus 5/- carr. £7/10/0

WE ALSO OFFER IT ASSEMBLED and READY FOR USE for

and READY FOR USE for (plus 5/- carr, and ins.)
This amplifier has proved one of the most popular models yet offered to the HOME CONSTRUCTOR. It provides really excellent reproduction up to 8 watts, employing 6V6's in push-pull and incorporating negative feedback. Provides for the use of both 3 and 15 ohm Spoakers. The Complete SPECIFICATION and BUILDING INSTRUCTIONS are available for 1/6.

"Wire-to-Wire" Diagrams are included and all Components are available separately.

SPECIAL PRICE REDUCTIONS . . . FOR PURCHASERS OF A COMPLETE "HI-FI" AMPLIFIER

WE WILL SUPPLY (a) COMPLETE KIT OF PARTS to build THE MULLARD "5-10" MAIN AMPLIPIER and the STERN'S "fidelity" PRE-AMPLIPIER-TONE CONTROL
UNIT for £16/16/- or we will apply THE TWO UNITS MADE UP and READY FOR USE for £19/19/s, Terms: Deposit £9/19/6 and 12 monthly payments of 18/7, or £5 Deposit and 9 monthly payments of £1/16/7.

RADIOGRAM" MODERNISE YOUR OLD

IT IS MUCH CHEAPER THIS WAY!!

THE LATEST DESIGN OF COMBINED AM/FM REPLACEMENT RADIOGRAM CHASSIS and a NEW 4-SPEED RECORD PLAYER STERN'S NEW "Fidelity" COMBINED AM/FM RADIOGRAM CHASSIS

A genuinely hand-made chassis providing really high quality on both Radio on both Radio on both Radio

PRICE £26/15/0 (Plus 7/6 carr. and ins.).

and ins.).
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Deposit £6/14/and 9 monthly
payments from
£2/9/-. H.P.
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and 12 monthly payments £1/4/10.

BRIEFLY IT HAS:-RRIEFLY IT HAS:—
An 8 valve line up incorporating the latest MULLARD preferred-type valves.
Provides complete coverage of the VHF/FM waveband plus the SHORT, MEDIUM and LONG waves.
Has ELS4's in Push-Pull, with negative feedback of 6 watte output.
Employs "Plano Key" Selector Switch and a Variable Tone Control.
Contains Gram input socket for both Crystal and Magnetic Pick-ups.
Provides for use of either 3 or 15 ohm Speakers.
Has "Magle Eye" Tuning Indicator.
Dimensions 13in. ×
9 in. × 8in. high, Dial size 11 in. × 5 in.

A genuinely hand-made chassis providing really high quality on both Radio and Gram.

A "de iuxe" Chassis for those who want the highest possible quality possible quality PRICE PRICE



£29/8/0

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TERMS: Credit
Deposit £7/7/and 9 monthly
payments of
£2/14/-. H.P.
Dep. £14/14/and 12 monthly
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£1/7/3.

BRIEF SPECIFICATION:—
A 9 valve line up cmploying the latest MULLARD preferredtype valves. Provides complete coverage of the VHF/FM
Transmissions, plus the SHORT, MEDIUM and LONG
waves. • Has Push-Pull output, with negative feedback,
for 6 watts Peak output. • Quilck action "Plano Key"
Selectors and separate Base and Treble Controls. • Has
"Magic Eye" Tuning Indicator. • Two Gram Inputs are
provided, one for Crystal Pick-ups and the other for Magnetic
types. • Dimensions ISIn. x 9\text{in. x 8ln. high, Dial size
11\text{in. x 5\text{in.}}

PRICE

£24/6/6

(Plus 7/6 carr. and ins.). TERMS: Credit Deposit \$6/2/0 and 9 monthly payments of \$2/4/7. H.P. Dep. \$12/3/3 and 12 monthly

and 12 monthly
payments of £1/2/7.
BRIEF SPECIFICATION:—
Full AVC on LONG, MEDIUM and SHORT wavebands and

Full AvC on LONG, MEDIUM and SHORT wavebands and amplitude limitation on FM bands. ⊕ Incorporates intermal aerial (Ferrite Rods) for local station reception on MEDIUM and LONG WAYES. ⊕ Coverage is FM 87 to 101 Mc/s, Short 16 to 50 metres, Medium 187 to 540 and Long Waveband 1000 to 2000 metres. ⊕ Recently developed Valve line np of ECCS5, ECHSI, EFS9, EABCS9, EL54, EZS0 and EMS0 Tuning Indicator. ⊕ Output is 4 watts and any 3 or 15 ohm P.M. Speaker can be used. ⊕ Connection Sockets on the Chassis (a) Pick-up, (b) Extension Speaker, for 3 or 15 ohm types. ⊕ Overall size 12in. × 9in. × 7‡in. high. Dial, 11‡in. × 6in.

I high. Dial, 114in. x 6in.

To semplete a GENUINELY HIGH QUALITY RADIOGRAM..., We recommend THE NEW COLLARO MODEL 456 4-speed Autochanger, and if a LOUDSPEAKER is required...

We recommend THE 8- or 10-inch We STENTORIAN "HIFF!" MODELS. We have SPECIALLY REDUCED PRICES for purchasers of a CHASSIS and RECORD PLAYER (and SPEAKER if required). SEED S.A.E. FOR DETAILS.

CASH ONLY OFFER!

This latest B.S.R. MONARCH 4-SPEED AUTOCHANGER £7/19/6 (Plus 5/- carr, and ins.). These units will autochange on all three speeds, 7ln., 10in, and 12 in.

They play
MIXED 7th.

10th. and 12th. records of same speed.

They have separate sapphires for L.P. and 78 r.p.m.,
which are moved into position by a single switch.

Minimum baseboard size required 14x124th., with
height above 54th., and height below baseboard 24th. A
bulk purchase enables us to offer these BRAND NEW
UNITS at this exceptional price.

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Phone: FLEet Street 5812-3-4.

RECORD **PLAYERS**

THE VERY LATEST MODELS ARE OFFERED AT GREATLY REDUCED PRICES

● TRANSCRIPTION UNITS. ● 3- and 4-SPEED AUTOCHANGER ● AUTOCHANGERS with MANUAL CONTROL POSITION. Send S.A.E. for ILLUSTRATED and DESCRIPTIVE LEAFLET.

SPECIAL CASH OFFER

A good quality 2 Stage (plus Rectifier) GRAM AMPLIFIER together with a 6in. P.M. Speaker and this attractive PORTABLE CASE.

ALL FOR ONLY 28/7/6. The Amplifier incorporates the latest B.V.A. Valves, types ECC33, EL84 with EZ80 Rectifier and has separate BASS and TREBLE CONTROLS. The CASE is attractively finished in Revine, maroon and grey, and has space for almost any make of Autochanger.

We also sell the two separately:
(a) AMPLIFIER and 64in. SPEAKER £4 12 (b) PORTABLE CARRYING CASE £3 17 6

FOR CALLERS ONLY
PARTS including F.M. Tuners, AM/FM Tuners, Midget
Battery Portable and Mains Units, etc., etc... We also have
the most comprehensive stock of WIRELESS and
ELECTRICAL COMPONENTS.

STERN'S 'SUPER SIX"

6 Valve Waveband Superhet

Provides good se-lection of stations and really good reproduction on both RADIO &

GRAM.

PRICE £14.0.0 (ptus 7/6 arriage and ins.

CREDIT THRMS: Deposit £3/10/- and 9 monthly payments of £1/5/8. H.P. TERMS: Deposit £7 and 10 monthly payments of 16/-.

BRIEF SPECIFICATION .

RIEF SPECIFICATION.

Delayed AVO on all wavebands
Preselection (seedback
Latest valve line-up: 12AH8, 6BA6, 6AT6, two 6AQ5s
and 574 (or equivalents)
Push-pull output gives approx, 6 watta
Connections on chassis for extension speaker, gram
and mains supply to gram.
Coverage 18-50 metres, 190-5802 and 800-2,000
Overall size 11x 7\$ x891m. high, dial 8\frac{1}{2} x491m.

A bronze dial escutcheon is available for 4/8.



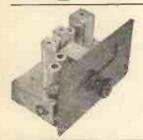
18. TOTTENHAM COURT ROAD, LONDON, W.1

MUSeum 5929/0095 (50 yards only from Tottenham Court Road Tube)

All post orders please to: -24-26, HAMPSTEAD RD., LONDON, N.W.1 EUSton 5533/4/5

LOOK ! ! ! A first-class SIGNAL GENERATOR that YOU can afford. The "Weyrad" yes S.G.M.L. covers 100 kc/s to 70 mc/s. in 6 bands, on Fundamentals. with an accuracy better than ±2%. Switched Audio Mod.: 500 c/s. For use on A.C. Mains 100/250 v. 50 cycles. Fully illustrated leaflet available. Price ONLY 212/10/- plus 2/6 P. & P. H.P. available.

ACOS TYPE 7 Crystal Microphone Inserts. Brand New, 7/8 each, post



THE JASON FM TUNER

Based on the booklet by Data Publications Ltd., 2/- post free, including our individually priced Parts List. Highly sensitive free from drift. Incorporates 4 valves 6AM6 and 2 specially graded G.E.C. Crystals. The kit supplied includes drilled chassis with tuning condenser, scale calibrated in m/cs., and attractive bronze stove-enamelled front plate already mounted (flustrated). Front plate size 8in. × 5in. chassis 7in. × 4½n. × 1½in. Complete standard kit £8/15/- plus 2/6 P. & P. Fringe area kit £7/15/-, plus P. & P.

FM POWER PACK KIT. We can now supply complete kit for power pack suitable for the above F.M. tuner or any other similar type. Price for the complete kit is 37/6 only, or 52/6 for ready assembled unit. This puck is extremely small, incorporating valve rectifier type 6X4 and built on chassis size only 6×4×1½in. Optional extra for power pack. Buigin Octal Plug 2/3.

THE T.S.L. FM TUNER!

THE T.S.L. FM TUNER!
We can now supply this FM/VHF adaptor either in kit form or fully assembled, wired and tested. Our price for the ready-built unit, which incorporates its own power supply is 21.3/15/- oily, tax paid, plus 5/. P. & P. or H.F. terms. Magic tye tuning indicator, just plug in, 19/- extra. Or the kit complete as specified 21.0/19/6 plus 3/6 P. & P. The booklet "FM TUNER CONSTRUCTION" (32 pages) with full technical data and point-to-point wiring diagrams together with our acparately priced parts list, is available at 2/6 post free.



THE T.S.L. AM/FM CHASSIS I

Exceptional value. Covers L.M. and S.W. plus FM: 8-valve push-pull output. Ferrite
rod acrial. Valve line-up ECC85, ECH81, EF89, EABC80, ECC82, two 6BW6s plus 5Y3
Large full-vision dial, size 14½ x6in. Chassis size, overall: 15 x7½ x8in. high. Tax paid,
26 guinesa, plus 5/- packing and carriage. A supplementary extra is magic eye EM34,
complete with escutcheon and fixing cable at 26/-. H P. terms available. Demonstrations
at 18 Tottenbarn Court Road!

at 18 Tottenham Court Road!

INTRODUCING THE NEW T.S.L. FM TUNER UNIT.

This compact unit with built-in power supplies has been designed by craftsmen to standards which will satisfy the most critical enthusiast. Brief specifications: Valve line-up: ECC85, 2-EF89, EABC80, 6X4 and EM80. Overall size: 10 jim. W. x5 jim. Hx 5jin. D. Dial size 10 jim. x x5 jim. Attractive plastic dial in Black and Gold with easy-to-read callbration. Controls: Bwttch. OFF FM & GRAM. and tuning. Pre-set gain control at rear of chassls. Connections: Co-Axlal output socket, 300 ohm aerial input socket and pick-up input socket on rear of chassls. Price £17/10/-, plus 5/- P. & P.

THE DULCI FM TUNER. Incorporates own power supply, suitable for use with any amplifier. Valve line-up. ECC85, two EF89, EABC80, 6X4 and EM80 indicator! Overall size: 9 x x 5 jim. injsh. Pre-Budget price £16/16/-, plus 5/- P. & P. Iliustrated leaflet available, also H.P. terms.

THE R.E.P. 1-VALVE RECEIVER. All-dry battery operation, for use with headphones The complete kit available at 42/-, less batteries plus 2/- P. & P. or full instructions at 9d. post free

COIL PACKS. Manufacturers' Surplus, Miniature size, only 2\square\text{in.} \times 2\square\text{in.} \times 1\square\text{in.} \times 2\square\text{in.} \times 2\square\text

TELEVISION TURBET TUNERS 12 CHANNEL—"TELENG"

We have six types now available from stock, to cover Bands I and III—fully Illustrated and descriptive leaflet available on request. Each unit is fully aligned and thoroughly tested before despatch. Valves employed are PCF80. PCC94 for AO/DC and ECF80 and ECC94 for AO.C. Price complete 27/7/-2/6 P. & P. Type. Sound MC/s Vision M/Cs Heater T7348 38.0 34.5 Series T734P 38.0 34.5 Parallel T7168 19.5 16.0 Series T7169 19.5 16.0 Series T7169 19.5 16.0 Parallel

TT:34P 38.0 34.5 Parallel TT:168 19.5 16.0 Series TT:169 19.5 16.0 Parallel TT:138 10.5 14.0 Series TT:13P 10.5 14.0 Parallel We have a large selection of in-built converters for all area from 92/6: also aerials low-loss co-axial cable at 9d. per yard Are you on our mailing list?

THE PAMOUS UNIVERTER—COM-PARE THE PRICE

Handsome wainut cabinet. Suitable all areas. Contains own power supply. Simply connect to aerial. Four-vaive circuit. Complete with all instructions. £6/19/8 plus 3/8 P. & P.

JUST ARRIVED!

The "Imperial" AM/FM Radiogram
Replacement Chassis. Overall measurements 13in. x 6in. x 7in. high. Dial
cut-out required only 104in. x 24in.
Covers long, medium and F.M. Features

include very attractive black and gold dial.
4-button, push
button unit.
gram position.
separate FM separate FM tuning, con-tinuously variable

tinuously variable tone control, A.C. mains 110/230 v. Valve line-up El-14. ECC85, EABC80, EF89, ECH81, plus metal rectifier. Price complete £16/19/6 tax paid. H.P. term: available.

DULCI AM/FM CHASSIS H4
Illustrated leaflet available. L., M. and
Short Waves plus F.M. This is a quality
chassis 6 latest B.V.A. Mullard Valves,
including magic eye. High Q inductances
throughout, also Ferrite rods. Price is
\$24/6/- cash—or H.P. terms.

SPECIAL PURCHASE! MANUFACTURER'S SURPLUS

SPECIAL PURCHASE! MANUFACTUSE
Owing to favourable purchase we can offer strictly
limited quantity of these bandsome chassis. AC/DC
200/250 v. for Medium and Long Waves, plus gran
position. Incorporates own frame serial. Valve
line-up: U107, N108, DH107, W107 and X109. Overlae
chassis size 12 x5 | x7 | in. high. Attractive bronze
dial with gold and cream lettering. Dial size 11; x 4 | in.
Scale length 7 | in. Logging scale provided.
Price 27/19/6 only, tax paid, plus 3/6 P. & P. H.P.
terms, £4 deposit plus four monthly payments of 22/-



Splendid unrepeatable offer!! Dulci F.3 Radiogram chassis

We have been very fortunate in being able to obtain a limited quantity of this well known and highly efficient chassis which we are able to offer at a greatly reduced price.

Specification:
Three waveband: long 1,000-2,000
metres: medium 187-540 metres; short
16-50 metres. Valve line-up: X79,
6BA6 or W747, 5A78 or DH77, EL84
or N709, 6X4 or U78.

Four controls, tone/on-off.

Volume wavechange.

Tuning: output 4 watts matched to 3-5 ohms. Incorporates latest Ferrite Rod Aerial. Pick-up sockets and mains supply for Gram motor. Overall dimensions: 12in. L. x 7 in. D. x 7 fün. H. Attractive dial with red, gold and green lettering on black background. Size 11 fm. x 4 jin. Price only £10/5/- plus 3/6 P. à P.

JUST RELEASED 111

THE NEW R.C. TRANSISTOR/CRYSTAL RECEIVER KIT

This receiver, covering medium waveband, which can be assembled in about 1 hour, will give amazing volume and tonal quality when used in conjunction with a good aerial and earth. Incorporating PNP Transistor and Germanium Diode. For headphone reception. Included with the kit of parts is a handsome plastic case in black and white, measuring 4½×2½×1½in. This case accommodates the complete receiver, including battery.

PRICE OF COMPLETE KIT: 25% plus 1/3 P. & P.

Lightweight high resistance headphones can be supplied separately at 15% pair. If, however, the kit is purchased complete with headphones the supplied at a SPECLAL INCLUSIVE PRICE OF 37/6 plus 1/6 P. & P.

Optional extra, 100t. coll single 7/36 coloured P.V.C. covered wire, suitable for both aerial and carth. 2/6 only.

TRANSISTORS 11 Now available, manufacturers surplus transistors. Suitable for use in Audio stages, etc., and for experimental purposes. P.N.P. type. Only 10%-cach, post free. Blue spot type available for B.F. up to 1.6 Me/s., 15%-each.

AM/FM KIT

Introducing the JASON AM/FM KIT for medium waves and F.M.

As illustrated this is a very high quality chassis incorporating 8 of the latest miniature valves, plus DM70 magic eye. Kits are available for chassis complete with output stage at £15.15/-. Also less output stage but with own bullt-in power pack at £13.19/6 only. These are high fidelity units and exceptional value at these prices which include all required components and full constructional details. Fully illustrated Data Booklet with full construction details, plus individually priced component list, available per return of post at 2/- post tree. Both plus 3/6 P. & P.

DULC! H4T AM/FM TUNER

This unit has been designed for Quality Reproduction and built to the highest technical standards. Contains own power supplies. Brief specifications: four wave-bands: VHF 87-101 mc/s. Short, 16-50 bands. VHF 87-101 mc/s. Short, 16:50 metres. Medium, 197-540 metres. Long. 1,000-2,000 metres. Intermediate Frequencies: 10.7 mc/s. and 471 Kc/s. Full A.G.O. on all A.M. Bands and amplitude limitation on F.M. Band. Three controls: ON/OFF, Volume, Tuning and Wave-bange. Valve line-up: ECCS5, ECH81. EF59, EABC90, EM90 and EZ80. Attractive dial in Black with Red, G.een and Gold lettering, size 11in. × 5in. Chassis size (overall) 12in. W. × 7in. H. × 8½n. D. A low impedance of 47K obm output makes matching to amplifier non-critical. Weight 12lb. £20/17/-, plus 5/- P. & P. Price 1s

THE R.C.3/4 WATT AMPLIFIER KIT Compare the advantages!

Treble bass. AND middle controls For crivatal or magnetic pick-upl A.C. Mains 200/250 v. Valve line-up: 6V6GT, 88G7, netal 6X5GT. Negative feedback, Built on stove enamelled steel chassis, measuring only 8ln, x din. x 1/1n. Four engraved cream knobs are included in the price of the complete kit with all necessary practical and theoretical diagrams at £4/5/- only, plus 2/6 packing and post or Instruction Book,

fully illus-trated for freel This amplifier can be supplied assembled, tested, and ready for ready ready for use at £5/5/plus P. ani
P. Hearing
is believing.



Our advantageous H.P. terms are avallable on any single item over £5. Let us have your

Please add postage under £1, or Cash with order. C.O.D. charge extra-open 9 a.m. to 6 p.m. Monday to Friday. Sorry but we close I p.m. on Saturday

THE "ECONOMY FOUR" T.R.F. KIT. A
three-valve plus metal recifier receiver. A.C.
mains 200/250 v. Medium and Long waves.
We can supply all required components right
down to the last nut and bott. Valve line-up
6K7, 637 and 6V8. Chassis ready drilled-Cabinet
size 12in, long by 6in. high by 5in. deep—Choice
of viory or brown Bakelite, or wooden walnut
finish cabinet. Complete instruction booklet
with practical and theoretical diagrams. Each
component brand new and tested prior to packing. Our price 25/10/- complete—Remember
this set is being demonstrated at our shop
premises! We proudly claim that our fully
illustrated instruction booklet is the most comprehensive available for this type of receiver—Booklet available at 1/6 post free,
allowed if kit is purchased later. Plus 2/6 packing and carriage for complete kit.

This is



cking and carriage for complete kit.

THE "SUPERIOR" FOUR KIT. Our new four-valve receiver. A.G. mains, 200/250 v. M. and Long waves. As with our very successful "Economy Four" all required components as supplied. Valve line-up: 2 6807, 8 X50T and 6 V60T. Chassis ready drilled. Cabinet size 10 in. x 10in. wide. Maximum depth at base 5in. tapering to 3 in. at top. Sloping front. Very attractively finished in light wainut and peach. Each component brand new and tested prior to packing. Complete instruction booklet with practical and theoretical diagrams is provided. Booklet available at 1/6 post free. Our price for complete kit. 26/9/61 Please add 2/6 packing and carriage. If preferred, we can call the complete for the post of the complete kit. 26/9/61 Please add 2/6 packing and carriage. If preferred, we can daily comprising and knobs. 4.56/-, plus 2/6 packing and carriage. N.B.—Our kits are even supplied with sufficient solder for the job.

N.B.—All our T.R.F. Kit circuits now include specially wound Denco "Max-Q" coils on polystyrene formers, improved performance! Price remains the same.

SURPLUS BARGAINS

METERS F.S.D.

50 microamp

50 microamp

50 microamp

100 microamp

200 microamp Size D.C. 2in. R.P. Price
R.P. 50/Rectangular 120/Rectangular 110/F.R. 45/F.R. (Tropicalised) 30/F.R. 18/8
F.R. 18/8
F.R. 17/8 D.C. 2in.
D.C. 5in.
D.C. 4in.
D.C. 3}in.
D.C. 2}in.
D.C. 2in.
D.C. 2in. 200 microamp 500 microamp F.R. 17/6
F. Sq. 22/6
F. Sq. (1954 manufacture by Eillott)
Deak Type 30/F. Sq. 8/6
F.R. 10/F. Sq. 6/6
F.R. 10/-1 mA. M.C. M.C. M.C. Thermo M.C. M.C. M.C. D.C. 21in. D.C. 21in. D.C. 21in. R.F. 21in. R.F. 21in. D.C. 21in. 1 mA. 50 mA. 500 mA. .5 amp. F. Sq. (shunt required)
B.P.
B.P. 10/-15/-45/-7/6 6/-1 amp. 120-0-120 amp. A.C. 4in. R.F. 2jin. R.F. 2in. 150 amp. 150 amp. 1 amp. 3 amp. 20 amp. 25 amp. 30 amp. 15 volt Thermo
 R.P.
 7/6

 F. Sq.
 6/

 R.P. (with shunt)
 10/6

 F.B.
 6/6

 F.B.
 12/6

 F.B.
 10/6

 F.B.
 12/6

 F.R.
 17/8

 F.B.
 35/

 F.R.
 30/ R.F. 2in. D.C. 2in. D.C. 2in. D.C. 2in. A.C. 2in. A.C. 2in. A.C. 2in. A.C. 3in. M.I. M.L. M.C. M.C. 00 volt M.L. 300 volt SPECIAL U.S. 0.1 mA. 24in. taken from equipment but perfect, 22/6 each. R.P. = Round Projection, M.C. = Moving Coll. Thermo = Thermo-coupled. F. Sq. = Flush Square. F.R. = Flush Round. M.I. = Moving Iron.
METER RECTIFIERS. 1 mA. by G.R.C., at 6/6, also 5 mA. by G.R.C. at 6/6.

METER SPECIAL. We have a limited quantity of aircraft electrical thermometers. Brand new, by Weston, 2in. moving coil meter, flush square fitting. These meters have a luminous scale graduated 40-140 degrees centigrade, but the full scale deflection is approximately 150 microsmps! Price 12/6 cach only, plus 1/- P. & P.

24 VOLT ROTARY CONVERTOR. Input 24 v. D.C. Output 200/250 v. A.C. 100
watts. Complete in black steel box
18in.x11in.x8in. Weight approx.
30ib. Completely smoothed, incorporates Sodium Lamp transformer Brand new, 92/6

HEADPHONES, Low resistance. Brand new ex.W.D. type CLR., 7/6 per pair, plus 1/6 p. 8 p. Brand new ex.W.D. type DHR, 4,000 obms, 15/- per pair. High resistance, 4,000 obms brand new light-weight. 15/- per pair plus 1/6 P. & P.

1/- P. & P.

TRANSISTORS! MULLARD TYPE O.C71.
Available ex stock at new list price of 24/- each also 0.0.71 at 21/- and 0.C.72
at 30/-. All post free.

No. 38 TRANSMITTER/RECEIVER
WALKER-PALKIE. Range approx. 5
miles. Coverage 7.4-9 Mc/s. The set only
complete with valves at 25/-, in very
good condition. We also have available
"38" sets absolutely complete with
junction box, serial headphonos, microphone, sparse set of valves, with full
operating instructions, absolutely brand
new in original sealed packing cases, at
99/6 each complete station, plus 5/carriage. Export enquiries invited.

SPECIAL PURCHASE from Ministry BRAND NEW No. 17 Mk. II TRANSMITTER/RECEIVER

Built into strong wooden cabinet 15in. x 14in. it into strong wooden capinet 13in, x 14in.

9in. Complete with headphones and
rophone. Range 5-8 miles with simple
ial. Frequency coverage: 44-61 mc/s,

(5-7 metres). Uses standard 120 v. H.T.
and 2 volt L.T. batteries. Complete with
full operating instructions. 59/6,
plus 3/6. C. & P. microphone.

No. 17 Mk. II, as above, but secondhand, in good condition and complete, 45/- plus 3/6 C. & P.

HEAVY DUTY VARIABLE AUTO TRANSFORMER. Input: Variable from 110 to 250 volts. OUT-PUT: 110 and 230 v. at 1.2 kVA. Contained in grey metal case with carrying handle. Overall size: 10tn. W. x 8tn. D. x 12in H. Weight 1 qtr. 19ib, Erand new! £8/10/-, plus 7/6 carriage and packing.

THE R.C. RAMBLER ALL-DRY PORTABLE

Full assembly details with practical and theoretical diagrams can be supplied at 1/6 post free. This is a truly professional 4-valve superhet—all dry—for medium and long waves A cream plastic top panel, with dial engraved in red and green adds to the very imposing appearance of this model which is housed in an attractive cream and grey leatherette covered attache-case type cabinet: measuring only 9ln. x 7ln. x 5 lin. Weight less batteries 4 lib., with batteries 4 lib. This set really has everything. Built-in frame aerial, high quality, extremely sensitive and very adequate volume from the 5 in. speaker. Valve line-up 3V4, 1R5, 185, 174. Also the required components, exactly as specified, including cabinet, can be supplied from stock at the special inclusive price of 27/7/- plus 2/6 p. and p. (less batteries). Uses Ever-Ready 90 v. H.T. type B128 at 10/-. Also L.T. 1.5 v. A.D. 35 at 1/8.



RAMBLER MAINS UNIT! At last we are able to offer our special mains units kit for using our popular all-dry "Rambler" on A.C. Mains. Complete kit, which when assembled fits snugly into batter-vecompartment, can be supplied at 47/6 plus 1/6 packing and postage. Price includes all required components, and full assembly instructions. N.B.—This unit is completely self-contained in a metal box measuring 7in. × 2½10. × 1½10. and is ideally suitable for ANY all-dry battery nortable requiring 90 y. H.T. and 1.5 L.T.

suitable for ANY all-drv battery nortable rea
THE R.C. 2 AMP. BATTERY CHARGER
KIT. Our new 2 amp. charger kit is now
available. Includes handsome wellentitled black stove-onamelled steel box.
size 7kin. × 3kin. × 3kin. Fully shrouded
first quality transformer, brand new G.E.C.
rectifier. Mains usa, etc., or charging 6 or
12 v. batt-ries at 2 amp. Absolutely complets kit with full practical and theoretical
instructions. Price 23/6 plus 2/6 P. & P.
Can be supplied assembled and tested at
41/6 plus P. & P. Heavy duty orcodils
clips suitable for car battery lugs. optional
extra at 1/6 per pair.
POWER PACK. By leading manufacturer.

POWER PACK. By leading manufacturer. Input 200/250 v. Output 380-0-380 280 mA, 6.3 v., 8 a., 6.3 v. 2 a., 4 v. 7 a., 5 v. 2 a., Fully smoothed. Incorporates valve reetifier G232. Chassis measures 13in. x 7in. x 5in. Wt. 22b. Few only at £4/19/6, plus 3/6 P. & P.

\$\frac{24}{19/6}\$, plus 3/6 P. & P.

8 v. VIBRATOR PACK RCA. Input 6 v.
Output 300 v. at 80 mA. D.O. fully
smoothed. Inco-porates 6-pin Vibrator,
024A Rectifier, heavy-dny transformer,
etc. Variable output. Dimensions
4x 44 x6ln. Brand new. Limited
quantity only, 35/x, plus 2/. P. & P.
6-VOLT VIBRATOR PACK. Ex-W.D.
6-voit input, output 140 v. 30 mA. Fully
smoothed and rectified, incorporating
6-void 4-pin vibrator. Unit size only
64 x5in. x2in. Price 15/-, plus 2/P. & P. New condition.

VIBRATOR PACK. Brand new, by Mallory, 12-voit input, 150 v. 40 mÅ output. Complete with synchronous vibrator, 12/6, plus 2/- P. & P.

FOUR-SPEED CHANGERS! The new B.S.R. 4-speed auto-changer in attractive cream and gold finish now available from stock at £9/15/- only, plus 3/6 p. and p. H.P. terms available.

GARRARD 120H | | | 3-speed, mixer, auto-changer unit with crystal t/o head and saphire styll. Provided with Manual control for single records. Finished in Cream and Brown. Band new in sealed manufacturers carton with fitting and operating instructions. 27/15/6 plus 5/- P. & P.

VALVES

VALVES

We have perhape the most up-to-date valve stocks in the trade. A stamp will bring complete list but the following is a selection only of brand new imported valve types. 'ully cuaranteed Purchase tax naid. EABCS0 10/- DAPSG 10/6 PLS3 11/6 EAHC 20/- DPSG 10/6 PVS1 10/- EB91 7/6 DKS9 10/6 PVS1 10/- EBCCS1 9/- DM70 9/- UFA1 10/6 ECCS1 9/- DM70 9/- UFA1 10/6 ECCS2 9/- DM70 9/- UFA1 10/6 ECCS3 9/- EMSG 11/6 UFA1 10/6 UFA1 10/6 ECCS3 10/- EWS9 11/6 EASCS 8/- ECCS9 11/6 EZS9 8/6 6AUG 8/6 ECLS0 11/6 EZS90 8/6 6BS9 9/- ECHS1 11/6 EZS90 8/6 6BS9 9/- EFS9 10/6 PCFS2 12/6 6AV 7/6 EFS8 12/6 PLS1 13/6 50B5 10/- EFS8 12/6 PLS1 13/6 50B5 1

In addition we naturally have an eurplus types available such as 6' etc. All in our valve price list! naual 6V6GT

INEXPENSIVE INTERCOMM! A pair of ex-W.D. high-grade sound powered brand new microphone/receivers and 15 yds. of twin wire. Simply connect for one-to-one working. No batteries required. Only 3/6 pr., plus 1/- P. & P. SUUND POWERED TELEPHONE HANDEETS. The luxury version of the above! No batteries required, simply connect with twin wire and use as telephone, 35/- per pair including 15 yds. flex, plus 2/6 P. & P. 100 yds. of twin flat transparent P.V.Q. flex, suitable for long distance, or all radio or electrical work, at 22/6 per coil, plus 1/- P. & P.

of electrical work, at 22/6 per con. plus 1/- P. & P.

MAINS TRANSPORMER HARGAINS.
Limited quantities. Manufacturer's Surplus 350-0-350, 80 mA., 6.3 v. 3 a., 5 v.
2 a. Half shrouded, drop-through, 14/6 only, plus 1/6 P. & P. 230 v. Input, 300-0-300 80 mA., 6.3 v. 3 a., 4 v. 2 a. Tropicalised drop-through type, 9/6 only, plus 1/6 P. & P. 10put 110/230 v. Auto load 230 v.
750 mA., 350-0-350 130 mA. Tapped filament winding 6 v. 3 a., 15 v. 3 a., 21.5 v. 6 a., also 5 v. 2 a. Tropicalised drop-through type, 21/- plus 2/6 P. & P. 270-0-270, 100 mA. 6.3 v. 3 a., 5 v. 2 a.
200/250 v. Input universal mounting, 16/6 plus 1/6 P. & P. Primary: 10-0-200-220-240v. Secondary windings: 375-200-0-200-375 v. at 250 mA. Double wound 230 v. at 6 amp., 6.3 v. at 3 amp. 8 v. at 0.25 amp. Unabrouded, fully impregnated drop-through type. 50/-, plus 1/6 P. & P. RECORD FLAYERS CABINETS—to suit all types of single record and autochanger units. Priced from 45/-. Send stam for fully illustrated list.

THE "R.C. STALLION." This is the latest addition to our range of gramophone amplifiers and is aupplied complete with high flux 8lin. P.M. speaker and baffle. Incorporating three octal-type valves, 607, 6V6 and 6X5, this robust and well-made unit is ideal for use in the larger type of record player or radiogram, and is equally suitable for use in the larger type of record player or radiogram, and is equally suitable for use in conjunction with a radio feeder unit. Separate bass and teeble controls are provided; also provision is made for an extension speaker and mains supplies to gram, motor Output approx. 4 watts. Size doveralle, mains 10/2 R. This amplifier will fit modification. Cabinet price 85/-, plus 6/6 P. & P. Will ...oo accommodate any standard Record Changer!

BUREAU CABINET. Handsome walnut fin is 8.

F r s n c b.

polished. MAINS TRANSFORMER BARGAINS.

finish, French polished. 33in. high × 31 in. × 15 in. Al-ready cut out for B.S.B. Mon-arch but sultable for RC54, etc. Uncut board

Oncut coard
for amplify
in g p or t i o n
measures 14\flash x 10\flash
Two record storage compartments.
The 215/15/- plus 15/- carr. H.P.
TERMS AVAILABLE.
We have a large selection of all type
cabinets. A stamp will bring list.

RADIO LT

18, Tottenham Court Road. London, W.I.

R. S. C.

ASSEMBLED CHARGER v. or 12 v. 2 amps. Fitted Ammeter and selector plug for 6 v. or 12 v. Louvred metal case, fin-ished attractive hammer blue. Ready for use with mains and output leads, Double Fused. 47/



BATTERY

ASSEMBLED CHARGERS
6 v. 1 amp. 19/9
6 v. or 12 v. 1 amp. 25/9
6 v. 2 amps. 29/9
6 v. or 12 v. 2 amps. 38/9
6 v. or 12 v. 4 amps. 56/9
Above ready for use. Carr. 29/9
HEAVY DUTY KIT

CHARGING

12 v. 30 amp. Suitable for garage or firm with a number garage or firm with a number of vehicles. Mains input 200/250 v. 50 c/s. Outputs 12 v. 15 amp. twice. Consists of Mains Trans. 2 Metal Rectifiers. 2 Meters. 4 Fuses. 4 Terminals. 2 Rheostats and circuit. Only 9 gns. carr. 15/-.

EQUIPMENT

BATTERY CHARGER KITS
Consisting of Mains Transformer, F.W. Bridge, Metal
Rectifier, well ventilated steel
case, Fuses, Fuse-holders,
Grommets, panels and circuit. Carr. 2/6 extra. 6 v. or 12 v. 1 amp.

Consisting of F.W. Bridge Rectifier 6/12 v. 5 a. Mains Trans., 0-9-15 v. 6 a. output, and variable charge rheostat with knob. Only 45/9. Post 3/- All for A.C. MAINS 200-250v., 50 c/s. Guaranteed 12 months.



Assembled 6 v. or 12 v. 4 amps.

Fitted Ammeter and Fitted Ammeter and variable charge selector. Also selector plug for 6 v. or 12 v. charging. Double fused. Well ventilated steel case with blue hammer finish.

Ready for see with mains and use with mains and output leads. Carr.

SELENIUM RECTIFIERS

2005141414	1/2 4 1 11 1 1 1 1 1
L.T. Types	6/12 v. 6 a 19/9
2/6 v. 1 a.h.w. 1/9	6/12 v. 10 a 25/9
6/12 v. 1 a.h.w. 2/9	H.T. Type H.W.
F.W. Bridge Types	120 v. 40 mA. 3/9
6/12 v. 1 a 4/11	250 v. 50 mA. 5/9
6/12 v. 2 a 8/9	250 v. 80 mA. 7/9
6/12 v. 3 a 11/9	250 v. 150 mA. 9/9
6/12 v. 4 a 14/9	300 v. 275 mA. 12/11
CO-AXIAL CABLE.	75 ohms, 1in., 8d. yard.
Twin screened feeder,	11d. yard.
5 CORE FLEX. I	Henleys circular rubber

14/36. Each lead colour coded. 1/6 yd. DIAL BULBS, M.E.S., 8 v. 0.2 a., 6/9 doz. 6.5 v. 0.3 a., 6/9 doz.

		current production	n).
1	OT E	x Govt.	
Tubular Typ	es	Can Type	
8 mfd. 450 v	1/9	16 mfd. 350 v.	1/11
8 mfd. 500 v	2/6	16μ F 450 v	2/9
16µF 350 v	2/3	16 mfd. 500 v.	3/9
16μF 450 v	2/9	32μF 350 v	2/11
		32 mfd. 450 v.	4/9
16μF 500 v	3/9	100 mfd. 450 v.	4/9
8-16μF 500 v.	4/11	8-8μF 450 v	2/9
25μF 25 v	1/3	8-16μF 450 V.	3/11
50μF 12 v	1/3	16-16μF 450 v.	3/11
50 mfd. 25 v	1/9	32-32μF 350 v.	4/9
		32-32µF 450 v.	5/9
50μF 50 v	1/9	64-120 mfd. 350)
100 mfd. 12 v.	1/9	v	7/6
100 mfd. 25 v.	2/3	100-200 mfd.	
6,000 mfd. 6 v.	3/11	275 v	6/11
Mo	ny othe	re in etock	

EX GOVT. VALVES. VR137, EA50, EB34, 11d; SP61 2/3; 4SHA 1/3; EL32 3/9; VS110 1/11; KT44 4/9; 6J5 3/9; 6V6C, 5U4G 5/9; 6K7G. 2/11; 35Z4, 6X4 5/9; EZ50, EF80 7/9.

EX GOVT. UNITS, type RDF1 in original sealed cartons with 14 valves including 5Z4G, etc., trans., L.F. choke, Rectifier, etc., etc. We cannot enter into correspondence regarding these units which represent a really exceptional bargain at 29/9. Carr. 7/6.

OIL FILLED BLOCK

CONDENSERS
Bryce 11-7 mfd. 500 v. New unused Govt. surplus, only 5/9 each.

THE SKY FOURT.R.F. RECEIVER



A design of a 3 valve 200-250 v. A.C. Mains L. & M. wave T.R.F. receiver with selenium rectifier. For inclusion cabinet illustra-

cabinet illustrated above or
walnut veneered type. It employs valves 6K7. SP61. 6F6G,
and is specially designed for simplicity in waring.
Sensitivity and quality is well up to standard.
Point-to-Point wiring diagrams, instructions
and parts list, 1/9. This receiver can be built
for a maximum of £4/19/6 including cabinet.
Available in brown or cream baleling are Available in brown or cream bakelite or veneered walnut.

EX GOVT. MAINS TRANSFORMERS

AII 230 V. 30 C/s. Hiput	
120-0-120 v. 40 mA	5/9
300-0-300 v. 150 mA., 4 v. 3 a	9/9
	- / -
250-0-250 v. 80 mA., 6.3 v. 3 a., 6.3 v. 1 a.	
	11/9
Potted 41-31-3in	11/7
	12/0
460 v. 200 mA., 6.3 v. 5 a	22/9

MANUFACTURERS SURPLUS
TRANSFORMERS
Fully shrouded upright. Primary 200-230-250 v. Sec. 425-0-425 v. 150 mA. 6.3 v. 3 a., 5 v. 3 a., 33/9. Clamped type 325-0-325 v. 100 mA., 6.3 v. 3.5 a., 5 v. 2 a. Wearite 19/9. 230-0-230 v. 60 mA., 6.3 v. 2.5 a. Midget, 2½ × 3×2½in. approx., 11/9. Post 2/9 on any of above.

EX GOVT. SMOOTHING CHOKES
250 mA., 5 H., 50 ohms 12
250 mA., 3 H., 50 ohms 8
150 mA., 10 H., 50 ohms 8
10 mA., 10 H., 100 ohms, Parmeko 100 mA., 10 H., 100 ohms, Tropicalised 3
150 mA., 50 H., 1,000 ohms
L.T. type 1 amp., 2 ohms 1 8/9

EX-GOVT. CASE. Well ventilated, black crackle finished, undrilled cover. Size 14—10—3\(\)\footnote{sins.} high IDEAL FOR BATTERY CHARGER OR INSTRUMENT CASE. COVER COULD BE USED FOR AMPLIFIER. Only 9/9, plus 2/9 post.

SPECIAL OFFERS. Small 2 gangs .0005 mfd., 4/9. Electrolyctics 32-32-32 mfd. 250 v. 2\(\)\footnote{structure} \times 1\(\)\footnote{structure} in lots of six, 2/3 ea.

TO MAINS CONVERSION R.S.C. BATTERY UNITS

Type BM1. An all dry battery eliminator. Size 5½ × 4½×2in. approx. Completely replaces batteries supplying 1.4 v. and 90 v. where A.C. mains 200-250 v. 50 c/s. is available. Suitable for all bettern with the supplying the supp available. Suitable for all battery portable receivers requiring 1.4 v. and 90 v. This includes latest low consumption types. Complete kit with diagrams 39/9, or ready for use, 46/9.



MINIATURE MOTORS. 24/28 v. D.C. or A.C. Size only 2½ × 1½ins. Spindle 1½ins. long, ½in. diam. Made by Hoover Ltd., Canada. Price only 9/9. VIBRATORS. Oak 2 v. 7 pin. synchronous, 7/9. M. E. SPEAKERS, '2—3 ohms R. A. 8in. Field 600 ohms, 11/9.

260-0-260 v. 70 mA., 6.3 v. 2 a., 5 v. 2 a. 350-0-350 v. 80 mA., 6.3 v. 2 a., 5 v. 2 a. 250-0-250 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a. 300-0-300 v. 100 mA., 6.3 v. 4 v. 4 a.,

Type BM2. Size 8 × 5½ × 2½in. Supplies 120 v., 90 v., and 60 v., 40 mA. and 2 v. 0.4 a. to 1 amp. fully smoothed. THEREBY COMPLETELY REPLACING BOTH H.T. BATTERIES AND L.T. 2v. ACCUMULATORS when corrected to LATORS when connected to A.C. mains supply 200-250 v, 50 c/s. SUITABLE FOR ALL BATTERY RECEIVERS normally using 2 v. accumulator. Complete kit with diagrams and instructions 49/9 or ready for use 59/6.

T.V. CABINETS. For 15, 16 or 17in. tube. Table model with doors, 79/6, carr. 7/6.

TRANCEORIA

ĺ	R.S.C. IKAI	M2 FORMER?
Į	FULLY QUARANTEED. INTER	LEAVED AND IMPREGNATED
ł	MAINS TRANSFORMERS	FILAMENT TRANSFORMERS
ı	Primaries 200-230-250 v. 50 c/s.	Primaries 200-250 v. 50 c/s.
ı	PHILLY SUPOLITION HADDIGED MOTIVATIVE	6.3 v. 1.5 a 5/9 0-2-4-5-6.3 v.
I	FULLY SHROUDED UPRIGHT MOUNTING	6.3 v. 2 a 7/6 4 a 16/9
ļ	250-0-250 v. 60 mA., 6.3 v. 2 a., 5 v. 2 a.,	0-4-6.3 v. 2 a. 7/9 6.3 v. 6 a 17/0 6.3 v. 3 a 8/11 12 v. 3 a. or
	Midget type, 2½-3-3in	12 v. 1 a 7/9 24 v. 1.5 a 17/0
	250-0-250 v. 100 mA., 6.3 v4 v. 4a., c.t.,	12 v. 1 d /// 24 v. 1.3 d 1//
	0-4-5 v. 3 a	CHARGER TRANSFORMERS
ı	250-0-250 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a. 23/9	All with 200-230-250 v. 50 c/s. Primarles
į	250-0-250 v. 100 mA., 6.3 v. 6 a., 5 v. 3 a.,	0-9-15 v. 11 a., 11/9; 0-9-15 v. 3 a., 16/9
	for R1355 conversion	0-3.5-9-17 v. 3 a., 17/9; 0-9-15 v. 5 a., 19/9
	300-0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a. 23/9	0-9-15 v. 6 a., 23/9.
	300-0-300 v. 100 mA., 6.3 v. 4 v. 4 a. c.t., 0-4-5 v. 3 a	ELIMINATOR TRANSFORMERS
	0-4-5 v. 3 a	Primaries 200-250 v. 50 c/s.
	300-0-300 v. 130 mA., 6.3 v. 4 a., c.t., 6.3 v.	120 v. 40 mA., 5-0-5 v. 1 a
	1 a., suitable for Mullard 510 Amplifier 33/9	90 v. 15 mA., 6-0-6 v., 250 mA 9/1
	350-0-350 v. 100 mA., 6.3 v4 v., 4 a., c.t.,	
	0-4-5 v. 3 a	OUTPUT TRANSFORMERS
	350-0-350 v. 150 mA., 6.3 v. 4 a., 5 v. 3 a. 33/9	Midget Battery Pentode 66:1 for 3S4, etc. 3/
	350-0-350 v. 150 mA., 6.3 v. 2 a., 6.3 v.	Small Pentode 5,000Ω to 3Ω
	2 a., 5 v. 3 a	Standard Pentode, $5,000\Omega$ to 3Ω 4/ Standard Pentode, $8,000\Omega$ to 3Ω 4/
	425-0-425 v. 200 mA., 6.3 v. 4 a., c.t.,	Standard Pentode, $8,000\Omega$ to 3Ω 4/ Multi-ratio 40 mA. 30:1, 45:1, 60:1,
	6.3 v. 4 a. c.t., 5 v. 3 a., suitable Williamson Amplifier, etc	90:1, Class B Push-Pull 5/
	450-0-450 v. 250 mA., 6.3 v. 6 a., 6.3 v.	Push-Pull 8 Watts 6V6 to 5 ohms 8/
	6 a., 5 v. 3 a	Push-Pull 10-12 Watts 6V6 to 3Ω or 15Ω 15/
	TOP SHROUDED DROP-THROUGH TYPE	Push-Pull 10-12 Watts to match 6V6 to
		2 5 9 150

16/9

22/9

29/9

36/6

26/9 23/9 33/9	ELIMINATOR TRANSFORMERS Primaries 200-250 v. 50 c/s. 120 v. 40 mA., 5-0-5 v. 1 a. 90 v. 15 mA., 6-0-6 v., 250 mA.	14/9 9/11
26/9 33/9 33/9	OUTPUT TRANSFORMERS Midget Battery Pentode 66:1 for 3S4, etc. Small Pentode 5,000Ω to 3Ω Standard Pentode, 5,000Ω to 3Ω	3/6 3/9 4/9

Small Pentode 5,000 Ω to 3 Ω	3/9
Standard Pentode, $5,000\Omega$ to 3Ω	4/9
Standard Pentode, $8,000\Omega$ to 3Ω	4/9
Multi-ratio 40 mA. 30:1, 45:1, 60:1,	-1-
90:1, Class B Push-Pull	5/6
Push-Pull 8 Watts 6V6 to 5 ohms	8/9
Push-Pull 10-12 Watts 6V6 to 3Ω or 15Ω	15/9
Push-Pull 10-12 Watts to match 6V6 to	,-
3-5-8 or 15Ω	16/9
Push-Pull 15-18 Watts, sectionally wound,	
6L6, KT66, etc., to 3 or 15 ohms	21/9
Push-Pull 20 Watt high-quality section-	
ally wound, 6L6, KT66, etc., to 3 or 150	47/9
Williamson type exact to spec	85/-
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I	SMOOTHING CHOKES	
l	250 mA., 5 H.; 100 ohms	11/9
ı	150 mA., 7-10 H., 250 ohms	11/9
Į	100 mA., 10 H., 200 ohms	8/9
ļ	80 mA., 10 H., 350 ohms	5/6
١	60 mA 10 H 400 ohms	4/11
	OU MARKET TO ARE TOO CHARLES	71 41

A6 ULTRA LINEAR 30 WATT AMPLIFIER R.S.C.



R.S.C. TA1 HIGH QUALITY TAPE DECK AMPLIFIER R.S.C. TAI HIGH QUALITY TAPE DECK AMPLIFIER FOR ALL DECKS WITH HIGH IMPEDANCE RECORD/PLAYBACK AND ERASE HEADS. Such as Lane, Truvox, etc., or matched to low impedance crase heads as inted latest COLLARO TAPE TRANSCRIPTOR. Chassis size 12:7-5in. Overall size 12:7-6ip. For 230:250 v. 60 c/cs. A.C. mains. Output for standard 2-3 ohm speaker. Ouly 15 millivoits input required for full recording, Ouly 2 millivoits minimum input required from recording head. Magic Eye recording level indicator. Provision for feeding P.A. amplifier. Can be used as gram. amplifier with input of 0.76 v. R.M.S. Negative feedback equalisation. Linear frequency response ± 3 D.B. 50-11,000 c/cs. Facilities for recordings at 16in. 7lin. or 3fin. Facilities for recordings at 16in. 7lin. or 3fin. per second. Automatic equalisation at the turn of a knob. When switching from record to heads is assured. PERFORMANCE IS COMPARABLE WITH UNITS AT OVER TWICE THE COST. LEAFLET 6d.

H.M.V. LONG PLAYING RECORD TURNTABLE COM-PLETE WITH ORYSTAL PICK-UP (SAPPHIRE STYLUS). Speed 33† r.p.m. BRAND NEW. CARTONED. Only £3/19/6 (Approx. half price). Carr. 5/- (for 200-250 v. A.C. Malns).

R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER TYPE A5



A highly sensitive 4-valve quality amplifier for the home, small club, etc. Only 50 millivolte input is required for full output so that it is suitable for use with the latest high-fidelity pick-up heads in addition to all other types of pick-ups and practically all mikes. Separate Base and Treble controls are revolted. These give full long playing record (qualisation. Hum level is negligible, being 71 D.B. down. 15 D.B. of negative feedback is used. H.T. of 300 v. 26 mA. and L.T. of 6.3 v. 1.5 a. is available for the supply of a Radio Feeder Unit or Tape Deck pre-amplifier. For A.C. mains input of 200-230-250 v., 50 c/s. Output for 2-3 ohm speaker. Chassis is not alive. Kit is complete in every detail and includes fully punched chassis (with baseplast) with the blue hammer finish, and point-point wiring diagrams and includes fully punched chassis (with baseplast), or assembled ready for use 25/extra, plus 3/6 carriage.

R.S.C. A7 3-4 WATT QUALITY AMPLIFIER

R.S.C. A7 3-4 WATT QUALITY AMPLIFIER
A highly sensitive 4-vaive amplifier using regative feedback
and having an excellent frequency response. Pre-amplifier
and Tone Control stages are incorporated with separate
Bass and Treble controls giving full tone compensation for
Long Playing records. Buttable for any kind of pick-up
including latest high fidelity types. H.T. of 250 v. 20 m.A.
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Unit, etc. ONLY 40 millivoits input required for full
output. Fully isolated chassis with baseplate. For A.O.
mains 200-250 v. 50 cycles, Output for 2-3 ohm speaker.
Complete kit of parts with point-to-point wiring diagrams
and instructions. Only £3/15/v. carr. 3/6 or factory built
22/6 extra.

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amplifiers.

Pleasey Ioin. 3 ohm with high flux density magnet. Only 28/9.

P.M. SPEAKERS. 2-3 ohm, 5in. Goodmans, 17/9, 7×4in. Elliptical, 19/6, 6jin. R.A., 15/9, 8in. Rola, 19/9, 10in. R.A., 26/9, 12in. Piessey, 29/11. 12in. Pissey 15 ohms, 10 watte 3 gms

GARRARD 3-SPEED AUTOMATIC RECORD CHANGERS.

Latest Model Mixer Type RC110. Fitted high fidelity turnover crystal pick-up head with dual point sapphire stylus. Baseboard size 14×124in., height above, 4in, Below, 24in. For 200-250 v. A.C. mains. Limited number. Brand new cartoned. Only £7/19/6, plus 3/6 carriage.

LINEAR 145 MINIATURE 4/5 W QUALITY AMPLIFIER, Suitable for use with Garrard, B.S.R. or any other record playing unit, and most microphones. Total negative feedback 12 db. Separate Jass and Treble Controls. For land the second of the second second of the second second

MICROPHONES. High fidelity crystal types. Acos 33-1 hand or desk type, 50/-. Plezzo with heavy floor base and telescopic stem, 26/19/8.

ROTARY CONVERTERS. 200 watts. Input 12 v. D.C. Output 230 v. 50 c/cs. A.C. Only 7 gns. Carr. 7/8.

PLESSEY DUAL CONCENTRIC 12 in. P.M. SPEAKERS



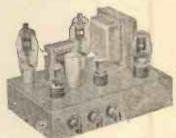
(15 ohms), consisting of a high quality 12in. speaker, of orthodox design supporting a small elliptical speaker ready wired with choke and condensers to act as tweeter. This high fidelity unit is highly recommended for use with our A8 or any similar amplifier. Rating is 10 watts. Price only £5/17/6.

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Terms C.W.O. or C.O.D. No C.O.D. under £1. Postage 1/9 extra on all orders under £2. 2/9 extra under £5 unless carriage charge stated. Full Price List 6d. Trade List 5d. 62, 2/9 extra under 65 unless carriage charge stated. Full l Open to Callers: 9 a.m. to 5.30 p.m. Saturday until l p.m.

R.S.C. ULTRA LINEAR 12-WATT AMPLIFIER



NEW 1956 MODEL AS HIGH-FIDELITY PUSH-PULL AMPLIFIER WITH "BUILT-IN" TONE CONTROL, PRE-AMP. STAGES

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High sensitivity. Includes 5 valves (897 outputs), High Quality sectionally wound output transformer, specially designed for Ultra Linear operation, and reliable small condensers of current magniacture. INDVIDUAL CONTROLS FOR BASS AND TREBLE "Lift" and "Cut." Frequency response ± 3db, 30-30,000 of 0s. Sit negative feedback loops. Hum level 71 db. down. ONLY 70 millivoits INPUT required for FULL OUTPUT. Suitable for use with all makes and types of pick-ups and practically all microphones Comparable with the very best designs. For STANDARD or LONG PLAYING RECORDS. For MUSICAL INSTRUMENTS such as STRING BASS, GUTTARS, etc. OUTPUT SOCKET with plug provides 300 v. 20 mA. and 6.3 v. 1.5 a. For supply of a RADIO FEEDER UNIT. Size approx. 12-9-7in. For A.C. mains 200-230-250 v. 50 c/cs. Output for 3 and 5 ohm speakers. Kit is complete to last aut. Chassis is fully punched. Full instructions and point-to-point wiring diagrams supplied.

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Unapproachable value at \$7/15/- or factory built 45/- extra. Carriage 10/-. If required louvred metal cover with 2 carrying handles can be supplied for 17/6. Where an extra input socket with associated volume control is required for mixing purposes this can be provided for 13/-extra. TERMS ON ASSEMBLED UNITS with extra input as mentioned above. DEPOSIT 25/6 and nine monthly payments of 23/4.

LINEAR "DIATONIG" 10-WATT HIGH PIDELITY AMPLIFICE. Incorporating pre-amp. For A.C. Mains input 200-230-250 v. 50 a.p.s. A compact attractively finished unit with two esparately controlled inputs, and outputs for 3 and 15 ohm speakers. Separate Bass and Treble controls. Five latest type miniature Mullard valves. Only 12 Gns. Carr. paid. Send S.A.E. for leaflet.

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SUPERHET RADIO FEEDER UNIT

SUPERHET RADIO FEEDER UNIT
Design of a high quality Radio Tuner Unit (specially suitable for use with any of our Amplifiers). A Triode Heptode Flehanger is used. Pentode I.F., and doube Diode Second Deteator. Delayed A.V.C. Ae./Grid Fl/C Coupling is by bottom end Condenser Coupling giving freedom from atignment rotubles when Ae. of varying lengths and capacity are used. Both Frequency Changers and I.F. valves are A.V.C. controlled from the very low distortion Double Diodes so arranged that very high Percentage modulation of the Transmitter can be handled without distortion. The Feed for the delayed A.V.C. is arranged so that A.V.C. distortion is avoided. The W. Ch. Sw. incorporase Gram. position. Controls are Tuning, W., Ch., and Vol. Output will load most Amplifiers requiring 500 M.V. input depending on Ae. location. Only 250 v. 15 u.A. H.T., and L.T. of 6.3 v. 1 amp, required from amplifier. Size of unit approx. 9-6-7in. high. Send S.A.E. for illustrated leaflet. Total building cost is 24,115. Point-to-point wiring diagrams and instructions. 2/6.

C.R.T. ISOLATION TRANSFORMERS

For Cathoas Ray Tubes having Heater/Cathode short circuit for C.R. Tubes with falling emission. Type A. Low leakage windings. Ratio 11.25 giving a 25% boost on Secondary. 2 vott 10/6 each With Tag

a 25% boost on Secondary.
2 voit 10/6 each 4 voit 10/6 each 10.8 voit 10/6 each 10/6 e

Output 2-2g-23-very- voits at 2 sinps. with lag raint 17/6 each.
All Isolation Transformers are individually boxed, labelled and clearly marked with relevant data.
NOTE:—It is essential to use mains primary types with T.V. receivers having series connected heaters.

RESISTORS. All values. 10 ohms to 10 meg., 1 w., 4d.;

I.F. TRANSFORMERS 7/6 pair

465 Ke/s Slug tuning Miniature Can. 21 × 21n. × 21n. High Q and good band width. By Pye Radio. Data sheet supplied.

HEATER TRANS. Tapped 200/250 v. 6.3 v. 11 amp., 7/6 COPPER PLATED AERIAL RODS. 1×12in. push fitting, COPPER PLATED AERIAL RODS. \$\(\pm\)2x12in. push fitting, 3/c oz. Post 1/c.

ALADDIN FORMERS and cores. Him. 8d.; \$\(\pm\)1n. 10d. 3in. FORMERS \$\(\pm\)83/8 and Cans TV1/2. \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)2x \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)2x \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)2x \$\(\pm\)1n. \$\(\pm\)2x \$\(\pm\)2

CRYSTAL MIKE INSERT by Acos Precision engineered. Size only 11 × 3/16in. Bargain. Price 6/6. No transformer required.

LOUDSPEAKERS P.M. 3 OHM. 2½m. square 17/6.

34 n. R.A. 17/6. 71n. × 4in. Goodmans 21/-,
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SWITCH CLEANER Finid, squirt spout, 4/3 in.
TWIN GANG CONDENSERS. 365 pl. midget, 1½n. × 1½n.
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SUPERHET COIL PACK 27/6 Ministure size 2½×2½×1½In. High Q dust cored coils. SHORT, MED., LONG. GRAM switching with connection disgram and circuit.

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SFEAKER FRET. Gold Cloth 18In. ×25in., 5/-. 25in. x
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5 p. 4-way 2 wafer, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
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3 p. 4-way, 1 p., 12-way, long spindle. 3/6
3 p. 4-way, 1 p., 12-way, long spindle. 3/6
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KNOBS, GOLD ENGRAYED. Wainut or Ivory. 1 in. diam., 1/8 each. "Focus." "Contrast," "Brilliance." Brilliance." "On-On." "Volume," "Voliton-On." "Toning." "Treble," "Bass." "Wavechange," "Badio Gram." "S.M.L. Gram." "Record-Play," "Brightness." ditto. not encraved. 1/-.



1957 RADIOGRAM CHASSIS
THREE WAVEBANDS
S.W. 16 m.—50 m.
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12 month Gnarantee. A.Q. 200/250 v. 4-way switch.
Short-Medium-Long-Gram. A.V.O. and Negative feedback. 4.2 watts, Ohassis 13½ in. x 5im. x 2½ in.
Glass Dial 10 x 4½ in., horizontal or vertical available. 2 Pilot Lamps. Four Knobs, Walnut or Ivory, aligned and calibrated. Chassis isolated from mains.
T.S.L. Tweeter supplied free.

BRAND NEW £10.10.0 Carr. 4/8. TERMS: Deposit £5/5/- and 6 monthly payments of £1.

AM/FM RADIOGRAM CHASSIS
Measurements 13in. × 6in. × 7in. high. Dial cut-out required only 10in. × 2in. 5 valve plus metal rect., gram. socket, plano key wavechange, tone control, med., long and V.H.F. wavebands. Valve line-up: ECC85, ECH81, EF89, EABC80, ELAI. For A.C. mains 100-250 v.

MATCHED SPEAKERS FOR ABOVE CHASSIS 8in., 19/6; 10in., 25/-; 12in., 30/-.

Collaro Autochanger RC531 for 78 R.P.M.
10in. and 12in. Records. Brand new in
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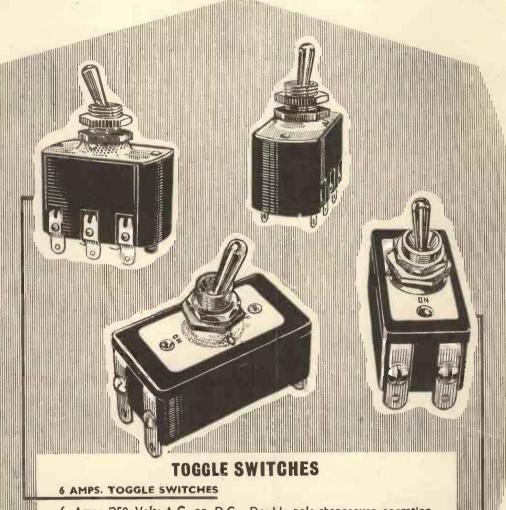
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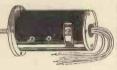
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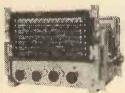
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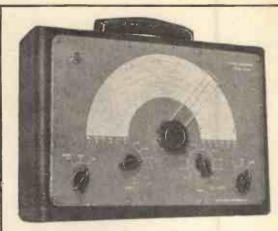
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TYPE 67

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Send S.A.E. for illustrated leaflet, or 1/3 for 14-page booklet which gives technical information, circuits etc. and is supplied.

which gives technical information, circuits, etc., and is supplied with each

Add carriage; 10/6 for Receiver, 5/- for Power Unit.

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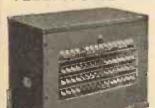
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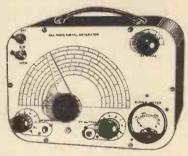
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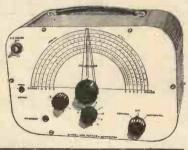


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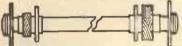
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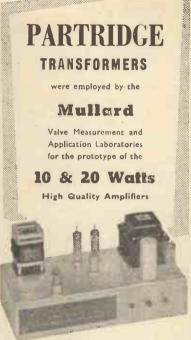
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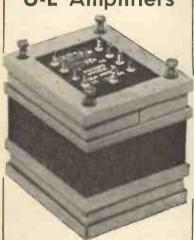
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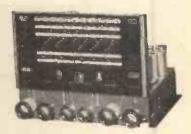
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and have reached a corresponding standard in English Language.

APPLICATION forms and further particulars may be obtained from the Departmental Train-ing Officer, Ministry of Supply, 66-72, Gover St., W.C.1. The closing date for receipt of completed application forms is Saturday, March 16, 1957.

REDERAL Government of Nigeria

March 16, 1957.

FEDERAL Government of Nigeria

SENIOR Telecommunications Engineers.
POST A.—To be responsible to a Principal Telecommunications Engineer for all engineering to a little of the Principal Telecommunications Engineer for all engineering the property of the Principal Telecommunications Engineer for all engineering at a district comprising two or more areas, including administration and control of engineering staff (African and European), tele-phone exchange systems (automatic and manual), 3 and 12 channel carrier systems on V.H.F. links, telegraph equipment and tele-printers, or to assist or deputise for the Principal Telecommunications Engineer at Territorial Headquarters.
POST B.—To be responsible to a Principal Telecommunications Engineer for the efficiency of all wireless communications including M.F. H.F. and V.H.F. installations in a major departmental territory and to plan and carry cut major overhauls and reconstruction of wireless installation and minor development work.
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WRITE Director of Recruitment, Colonial Office, London, S.W.I. giving age, qualifications and experience—posts and Telegraphs Department, Fiji.

Habito engineer—posts and Telegraphs Department, Fiji.

Rabito engineer—posts and Telegraphs Department, Fiji.

The responsible for the preparation of return be responsible for the preparation of return for the responsible for the preparation of return for the responsible for the responsible for the responsible for the responsible for the responsibility of the responsibil

SERVICE. Service.

CANDIDATES must be A.M.I.E.E. and be thoroughly conversant with the installation, operation and maintenance of low powered H.F. and V.H.F. radio stations and have a sound knowledge of the International Radio Regulations and procedure regarding frequency registration.

registration.
WRITE Director of Recruitment, Coionial
Office, London. S.W.1, giving age, qualifications and experience, quoting BCD 108/49/02 RADIO Maintenance Technician required by

POLICE Department, Government of Northern Rhodesia, for one tour of 36 months in first instance; salary, according to age and experience, in scale £705 rising to £1,200 a year; free passages; liberal leave on full salary; candidates, preferably aged 25 to 30, must possess academic qualifications in mathematics and physics of matriculation standard, together with sound knowledge of installation and maintenance of modern low and medium powered V.H.F. static and mobile equipment, H.F. transmitters and receivers, petrol generators and diesel electric sets.—Write to the Crown Agents, 4, Millbank, London, S.W.I. State age, name in block letters, full qualifications and experience and quote M2C/41913/W.F.

CHIEF Engineer reculred London area, to take charge of developing organization dealing with industrial electronic equipment, electronic components and printed circuits, this appointment demands an engineer with wide knowledge of industrial electronics; of manufacturing methods, both mass production and custom building; and with exceptional drive and organizing sbillty, age preferably 35 to 45; applications should give very full particulars of education and experience and will be treated in strict confidence.—Box 5690. [6663]

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Departmental territory and be responsible to a Principal Telecommunications Engineer for the efficiency of all wireless communications in the territory of all wireless communications in the territory of all wireless communications in the territory. POST B.—Will be in charge of all Posts and Telegraphs Department activities in an area, including control of staff, telephone exchange control systems (auto and manual) three and twelve channel carrier systems on open wire line, or VHF links and VF telegraph equipment and teleprinters. He will also be responsible to a Senior Telecommunications Engineer for the maintenance of all telegraph and telephone services in his area.

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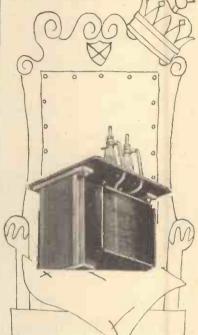
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W.I. [0262]
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