Wireless World June 1968 Three Shillings

W.W. colour receiver Stereo loudness control Keyswitch is proud to announce five new miniature relays that cleanly sweep the field. British designed, British developed, and British made, the KMK range features high contact capacity, moulded assembly for high insulation, phosphor bronze contact springs, 99.9% silver or silver cadmium oxide contacts, Swedish iron magnetic circuit, international contact clearance of 4mm, life in excess of 5 million operations, connection by solder or push-on type '110' connectors, open relays mounting in any position, and plug-in relays for international plugability. Unit prices are as low as 9/4d (1,000 rate), substantially less for larger quantities. These versatile new midgets are available with one, two or three

changeover contacts rated up to 10A at 250Vac/6Vdc, and with coils for 6-230Vac/2.5VA and for 6-200Vdc/1W. Contact Keyswitch for complete price and technical details of these exciting new all-British KMK's.

Keyswitch Relays Ltd, Cricklewood Lane, London NW2; telephone: 01-452 3344; telex: 262754.

1

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what has changed?

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And pickups, too. The trend here is towards smaller and lighter moving parts producing lower outputs, requiring greater sensitivity and improved signal to noise ratio in the pre-amplifier.

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Wireless World, June 1968

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7

Custom built amplifier installations from standard modules



Philips introduce versatile audio system.

Philips make it very easy for you to realise your own ideas for an audio system, by introducing an exclusive new formula based on two principles: modular construction and stacking ability.

Our modules are designed with common dimensions, styling and controls and are easily stacked.

The resulting high quality system can be extended when the need arises by stacking additional items. Philips modules enable you to meet your plans exactly, without compromise — efficiently and economically. Use them individually or stacked; free standing or wall mounted and where required, in 19 inch racks. Cases are finished in dark grey which, combined with the lighter grey of the panels, achieve a pleasant two-tone effect.

The modular range includes: 25 W, 50 W and 100 W power amplifiers (stackable to provide higher outputs if required); preset, mixing, and push button pre-amplifiers (with remote control facilities) which fit into each of the power amplifiers; an AM/FM radio tuner; a record player and an automatic tape recorder. The range is completed by a series of complementary accessory units, including 19 inch rackmounting panels.

Intermatching and other problems common to conventional equipment are completely eliminated. Choose your input, control and output requirements, then make the combination. Installations are ready for immediate use, suitable for continuous operation, reliable.

Use the Philips stacking system and obtain the exact audio system required for any indoor or outdoor application.

Equipment survey of the Philips stackable audio module system

The basic components of the system are housed in attractively styled standardised modular cabinets, finished in dark grey with light grey fronts. Cabinet width is 400 mm, depth 350 mm. The high quality equipment is suitable for all audio applications and for continuous operation. Reliable solid state circuitry throughout. Complete coverage of the audible spectrum at extremely low distortion.

Power amplifiers

Fully transistorised high quality power

installations. Mains powered. Easy selection of output voltage by means of movable sockets at the rear of the cabinet. Built-in limiter with pushbutton control on front panel. Input from various interchangeable preamplifiers which fit into the housing. Up to five power amplifiers may be fed from one pre-amplifier. Mains voltages: 110, 127, 220, 245 V. Distortion: $1.5^{\circ}/_{\circ}$ at nominal output. Output voltage: loudspeaker output 100, 70, 50, 35, 25, 10 V; line output 4 V. Sensitivity: 100 mV adjustable. DC supply: a DC voltage is supplied by the power amplifier to feed the pre-amplifier. Height: 115 mm, stacked 105 mm.

amplifiers for use in all audio

25 W power amplifier LBB 1001 output power: 25 W. power consumption: no load 16 W,

full load 75 W. 25 W mixing amplifier LBB 1021 as LBB 1001 with mixing pre-amplifier

LBB 1020. 50 W power amplifier LBB 1002

output power: 50 W.

power consumption: no load 41 W, full load 153 W.

50 W mixing amplifier LBB 1022 as LBB 1002 with mixing pre-amplifier LBB 1020.

100 W power amplifier LBB 1003 output power: 100 W.

power consumption: no load 33 W, full load 283 W. 100 W mixing amplifier LBB 1023

as LBB 1003 with mixing pre-amplifier LBB 1020.

Pre-amplifiers

Pre-amplifiers are special units which fit into each of the basic power amplifiers. Controls are grouped on a panel, which replaces a blind section of the power amplifiers front. Supply voltages are obtained from the power amplifier. Distortion: $0.5^{\circ}/_{\circ}$ (40 . . . 20 000 Hz).

preset pre-amplifier LBB 1010 Inputs: 4 microphone channels, sensitivity 0.3 mV. 1 channel, sensitivity 100 mV for use with a crystal pick-up, radiotuner or tape recorder. Presettings: separate gain adjustments per input channel. Front control: common gain control —

fader type — for the overall programme level.

LBB 1020 mixing pre-amplifier Inputs: 2 microphone channels, 2 channels adaptable as microphone channels or as input channels for crystal or dynamic pick up, by using microphone insert LBB 1025 or gramophone insert LBB 1026. Presettings: separate gain control per microphone input channel. Front controls: 4 separate gain controls — fader type — for each of the input

channels, 1 common bass control - fader type 1 common treble control -- fader type 2 separate bass cut switches for each of the microphone channels. Screw-in microphone insert LBB 1025 to extend the input possibilities of mixing pre-amplifier LBB 1020. Screw-in gramophone insert LBB 1026 to extend the input possibilities of mixing pre-amplifier LBB 1020. Push-button pre-amplifier LBB 1030 Inputs: 4 universal channels, one of which has priority function, max. sensitivity 0.3 mV. Presettings: sensitivity per channel, bass cut per channel, gain per channel, common bass and treble adjustments. Front controls: 4 independent pushbuttons for channel selection and 1 common cancellation push-button; selection buttons are illuminated when operated; channel 1 has priority e.g. for emergency; 1 common gain control fader type — for channels 2, 3 and 4. Remote control: The pre-amplifier has connections for remote control and

Power amplifier



signalling on all channels.

... with mixing pre-amplifier



... with push-button pre-amplifier





Tuner



Record player



Recorder



Tuner, record player and recorder equipment

radio tuner LBB 1037 Consisting of the high quality radio tuning unit GH 924, built into a standard stackable cabinet. Frequency bands: 150 . . . 400 KHz (long wave) 517 . . . 1622 KHz (medium wave) 5.9 . . . 18.2 MHz (short wave) 87.5 . . . 108 MHz (FM band) (the FM section is equipped with stereo facilities) Height: 150 mm, stacked 140 mm. record player LBB 1038 Consisting of the high quality record player GC 030 built into a stackable cabinet. Drawer type chassis and tinted methacrylate front. The module cabinet is also available separately with a blank chassis, suitable for mounting any other record player. Chassis is self-locking when pushed into the cabinet.

Height: 220 mm, stacked 210 mm. tape recorder LBB 1039 Consisting of the high quality tape recorder EL 3572, built into a stackable cabinet. The recorder is

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mounted on a drawer type chassis which is self locking when pushed into the cabinet.

Height: 220 mm, stacked 210 mm. Accessories

cable transformers LBC 1100 Completely mu-metal screened 500 Ω / 500 Ω high quality microphone transformers for isolating or balancing purposes. Mounting sets LBB 1045 Two threaded rods of 140 cm length, two strips, four nuts and four washers. Rods can be cut to length after mounting. 19 inch rack mounting panels LBB 1035 and LBB 1036 For mounting stackable modules in existing 19 inch systems.

Mounting hinges LBB 1046

For mounting a complete stack to a wall.

Electro-acoustics Division of Philips Industries, N.V. Philips' Gloeilampenfabrieken, Eindhoven, The Netherlands.



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• 9 A.C. and 9 D.C. voltage ranges from 150 millivolts to 1500 volts full scale • 7 resistance ranges from 150 millivolts to 1500 volts full scale • 7 resistance ranges, 10 ohms centre scale with multipliers $\times 1$, $\times 10$, $\times 100$, $\times 1k$, $\times 10k$, $\times 100k$, and $\times 1$ meg... measures from one ohm to 1000 megohms • 11 current ranges from 15 μ A full scale to 1.5A full scale • 11 megohm input impedance on D.C. • 10 megohm input impedance on A.C. • A.C. response to 100 kHz • 6in. 200 μ A meter with zero-centre scales for positive and negative voltage measurements without switching • Internal battery power or 120/240 volt A.C., 50-60 Hz • Circuit board construction for extra-rugged durability.

New Solid-State Volt-Ohm Meter, IM-16

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15

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MU.7522	3.75/15*	1-3, 2-4	100K.	6-8	82:1/164:1	Low Z. Mic/Grid
MU.7523	75/300*	1-3, 2-4	600 (C.T.)	6-7-8	1.41:1/2.82:1	Line/Line
MU.7524	150/600*	1-3, 2-4	600 (C.T.)	6-7-8	1:1/2:1	Mixing :Bal./Unbal.
MU.7525	600 (C.T.)	6-7-8	300/1·2K*	1-3, 2-4	1+1:1·41 (C.T.)	Mixing : Hybrid‡
MU.7526	600 (C.T.)	6-7-8	2·5k/10k.*	1-3, 2-4	2.04:1/4.08:1	Line/Grid
MU.7527	150/600*	1-3, 2-4	100K.	6-8	13:1/26:1	Line/Grid
MU.7528	7.5/30*	1-3, 2-4	600 (C.T.)	6-7-8	4·47 :1/8·94 :1	Low Z. Mic./Line
MU.7529	50/200*	1-3, 2-4	600 (C.T.)	6-7-8	1.73:1/3.46:1	Mic. or Line/Line
MU.7530	10K. (C.T.)	6-7-8	10K.	1-4	1 (C.T.) :1	600 Line Bridging
MU.7532	7.5/30*	1-3, 2-4	100K.	6-8	58:1/116:1	Low Z. Mic./Grid
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Wireless World

Electronics, Television, Radio, Audio

Fifty-eighth year of publication

June 1968

Volume 74 Number 1392



This month's cover. A development engineer at Mullard's Mitcham works adjusts a cooled two-stage parametric amplifier designed for use as a low-noise r.f. input stage in communications satellite earth stations. The amplifier operates at 4GHz with a bandwidth of 500MHz, and has a gain of 40dB and an overall noise temperature of $20^{\circ}K$.

Iliffe Technical Publications Ltd., Managing Director: Kenneth Tett Editorial Director: George H. Mansell Dorset House, Stamford Street, London, SE1

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CED 55
Wireless World

Constructional Projects

Although the final analysis of our recent reader interest survey is not yet available—it was essential to await replies from overseas readers who, incidentally, number one in five—one thing is very obvious from the comments and accompanying letters received: there is still a tremendous interest in constructional projects. It is, however, necessary to qualify this interest as readers made it blatantly plain that they do not want gimmicky electronic gadgets but worthwhile projects. Wireless World has for more than 40 years had a very strong following of 'do-it-yourself' enthusiasts, and as older readers will remember there were few issues, even when weekly, that did not include details of a receiver, or some item for home construction. In those far-away days, of course, it was cheaper to make than to buy a ready made receiver. Although this is, in general, no longer true the interest in constructional projects has continued and is by no means confined to those of our readers who have a purely amateur interest. This is amply borne out by the fact that professional readers frequently make use-of W.W. designs in professional equipment —and sometimes they appear on the market as commercial products!

While readers have applauded the inclusion of such designs as the Dinsdale and Bailey amplifiers, some have complained of the paucity of designs 'with the hall mark of the *Wireless World* laboratory'. In self defence we must point out that, whereas in industry a whole team of engineers will devote months or even years to the development of a project, we with a small editorial team (whose prime function is to produce a journal every month) are also expected to develop sophisticated pieces of equipment and then to describe them in sufficient detail to enable readers to build them. It will be obvious therefore, that the number of *W.W.* designs must inevitably be very limited.

For many months past W. T. Cocking, who was responsible for the design of the *Wireless World* monochrome television receiver published in 1947, has been perfecting constructional details of a colour receiver, and the first of a series of articles will be found elsewhere in this issue. However, we would stress that this project 'is not by any to be enterprised or taken in hand, unadvisedly, lightly or wantonly' (to quote from the Church of England marriage service). Having said that we believe that there will be very many readers who will enjoy the exercise. We would especially commend it to the heads of science departments of schools and colleges who would, we feel sure, find in it a worthwhile project which should have more than academic interest for the students.

In saying this about W.W designs we would not wish it to be thought that we belittle, in any way, the admirable designs submitted by contributors and we are in fact collaborating with some of them to bring to fruition more speedily ideas which we feel would be of particular interest.

One other aspect of circuit design must be mentioned. Although comparatively few articles are published in which full constructional details are given we know that the design data given in many others is used by readers to produce their own equipment (see note on p. 182). It would be true to say that rarely is there an issue in which there is not something for the engineer who wants to 'have a go' at constructing a piece of equipment for some particular aspect of the fragmentary technology we call electronics.

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Wireless World Colour Television Receiver

1. General considerations

This article is the first of a series describing a colour television receiver in detail. The receiver is one which has been developed by W. T. Cocking in the *Wireless World* Laboratory and is based upon Mullard design data. It is naturally much more complicated and expensive than a purely monochrome set, and its proper adjustment is not only more difficult but demands the use of more elaborate test equipment.

It must be emphasised at the start that the construction of a colour television receiver is no easy task. Apart altogether from the magnitude of the constructional work, a good deal of practical experience with black-and-white equipment is essential. The constructor must also possess, or have access to, quite a lot of measuring and test equipment. He must also, of course, understand the principles of colour television. In this connection, the series of articles by T. D. Towers in the January to December 1967 issues of *Wireless World* must be regarded as compulsory prior reading.

The final anode of the cathode-ray tube of a colour set normally operates at 25kV. High-voltage insulation problems arise, therefore, and certain safety precautions must be taken. Because of the high voltage, X-rays are generated in one valve and, to a smaller degree, in the c.r.t. itself. Since Xrays can be a danger to health, it is necessary to include proper screening against them. It must be strongly emphasised that the constructor must exercise proper care in the construction and use of the equipment and must never take short cuts by omitting safety precautions.

No danger arises in the normal use of the receiver. It is in initial setting up and in fault-finding that the possibility of danger arises, simply because one is then very close to the apparatus.

A further difficulty which may confront the intending constructor is that of obtaining some of the components. It is not possible to construct a colour receiver using only parts which are readily available on the retail market. Apart from the colour tube itself, the scanning coils and transformers, the convergence coils and some control coils for them are very definitely 'special to colour'. They are primarily manufactured for, and supplied to, the manufacturers of colour television receivers and their makers do not normally supply them direct to the public. It is hoped that some retailers will make supplies available, but it must be made clear that Wireless World can do little or nothing to help any individual who finds difficulty in obtaining parts. Details of suppliers will, of course, be given where possible. Before purchasing any components, therefore, it is wise to make sure that one can obtain *all* special parts.

The purpose of describing the construction of a colour receiver is primarily educational. It is envisaged that those who build it will be mainly people who in one way or another are engaged in the electronics industry and wish to gain experience with colour television. There is no reason why the amateur who is knowledgeable in television and who has already successfully constructed black-and-white receivers should not attempt it, although he will be doing something more difficult than he has previously tackled. Most definitely, however, it is not for the complete beginner in television.

All this preamble may sound rather discouraging. In fact, it is deliberately intended to discourage those who have not sufficient technical knowledge and proper facilities from attempting it. The total cost of the parts on the retail market is likely to be more than that of a commercial colour set and so the cost of failure to an individual is likely to be important. There are so many parts involved that there is a high probability of finding at least several faulty ones, and of making several wrong connections in the wiring! When one first switches on the completed set, therefore, it will be with the knowledge that there are quite likely to be half a dozen definite faults which will have to be traced before anything approaching good results can be obtained. This is very good educationally, for there is nothing like fault-finding to learn about equipment!

A general block diagram of the colour receiver is shown in Fig. 1. Without the block marked Chrominance Circuits, the equipment will produce a monochrome picture of any colour. Up to the video detector the circuits are conventional, but one has to be rather more fussy about bandwidth than in the ordinary black-andwhite set. The video amplifier is more complicated, partly because a 0.6-µsec delay line must be included to equalize the transmission times of the luminance and chrominance channels. In the main, however, it is more complex because it must provide a greater signal output than the ordinary video amplifier. The colour tube needs a video signal drive at the cathode of some 140 volts peak white to sync pulse. This signal is applied directly to the cathode of the red gun and through frequency-compensated potentiometers to the green and blue cathodes, since it is necessary to adjust the drives to the three guns to compensate for variations in individual tubes.

Fundamentally the timebases are the same as those of a black-and-white set, but they operate at higher power because they have to provide much larger scanning currents. This comes about because of the 25-kV final anode voltage of the tube and, in its turn, this is needed because of a basic inefficiency in a shadow-mask tube. The shadow mask intercepts a large part of the beam current of all three guns, current which in a normal tube would be focused on the screen to excite it. As it is, with a maximum beam current of 1.2mA at 25kV the e.h.t. power reaches no less than 30W! It is essential, too, that the voltage be stabilized and the usual practice is to employ a valve connected in shunt with the tube to absorb any current not needed by the tube. When the picture is black and the tube draws no current, the stabilizer takes the full 1.2mA and so dissipates 30W.

Because the tube operates at 25kV, the deflection volt-amperes needed are large. In one sense this is fortunate, since it enables the e.h.t. supply to be generated in the usual way as a by-product of the line scanning; it does mean, however, a high-power line-scan stage. The output valve, for instance, has to pass a peak current of 0.9A at an h.t. supply line of 285V. The mean current is 0.36A so the power input to the line-scan output stage is over 100W. Ventilation problems arise, therefore.

A common commercial practice is to obtain the e.h.t. supply from a valve rectifier fed from an overwind on the line-scan transformer. One disadvantage of this is that it provides another source of X-rays. These are produced whenever electrons strike a target, such as the anode of a valve, with sufficient velocity. X-ray production is usually considered to start at around 15-16kV, so it may occur in some degree even in monochrome sets. It rises rapidly with voltage, however, and at 25kV is quite considerable.

A valve rectifier is not used for the e.h.t. supply in the Wireless World receiver, but instead a voltage-multiplier with selenium rectifiers. This completely eliminates X-ray production at this stage. It is still present in the voltage stabilizer, however, and so it is necessary to screen this valve against X-rays. The colour tube itself can generate X-rays because it operates at 25kV. Lead glass is used in its construction and there is an external metal screen covering a large part of the bulb, and these cut down the radiation considerably. In the normal operational use of a colour set there is believed to be no health risk from X-rays at all. There may be some slight risk when one is very close to a set, as one must be in fault-finding, and, to a lesser extent, in adjusting it. The danger lies in total dosage which depends on the product of X-ray intensity and time. The writer considers it would be unwise to work all day



Fig. 1. Block diagram of the colour television receiver. The blocks do not necessarily correspond to physical units. Usually several blocks are contained in a constructional unit.

and every day close to the 'works' of a colour set of this type. This is not necessary, however. Provision is made in the design for the ready disconnection of the h.t. supply to the line timebase. Most fault-finding in circuits other than the line timebase and e.h.t. supply can thus take place with the X-ray producing stages inoperative. Provision is also made for operating the line timebase at reduced output at which the e.h.t. supply is 15kV only and X-ray production is negligible. Under this condition much fault-finding on the line timebase itself can take place in complete safety even if it takes a long time.

To revert to the block diagram of Fig. 1. the main difference from a black-and-white receiver lies in the convergence circuitry. This is necessary to obtain proper superposition of the three pictures produced by the three guns of the tube. The three guns in the tube cannot be situated physically at the same place and it is necessary to deflect their beams in such a way that their deflection by the scanning coils appears to take place from a common centre. Mounted on the tube neck between the deflector coils and the guns is the convergence assembly which comprises six coils and five permanent magnets. Three of the latter are static convergence magnets, the other two are 'purity' magnets. In addition, a further coil and magnet are mounted behind the main convergence assembly; this is usually called the blue-lateral assembly.

The purity magnets are two metal rings transversely magnetized to produce a field across the neck of the tube. The strength of the field can be controlled by rotating one magnet relative to the other; in this way, their fields can oppose or assist each other in any required degree. The direction of the field across the tube neck can be varied by rotating both magnets together. The purity magnets are normally adjusted with the red gun operative only, and the deflector coil assembly pushed back on the tube neck, to obtain the most pure red in the centre of the screen. The deflector coils are then moved forwards to obtain a uniform red over the whole screen. The purity is then checked for the other colours by putting the other guns on by themselves one at a time.

Convergence

The other four permanent magnets are next adjusted to secure convergence at the centre of the screen. A cross-hatch pattern generator is almost essential for this. The blue gun is turned off and two magnets are adjusted to superimpose the red and green bars of the pattern at the centre of the screen to produce a yellow pattern. The blue gun is then turned on and the remaining two magnets adjusted to superimpose the blue and yellow patterns in the centre to produce white.

These static convergence adjustments are quite easy to carry out, but they give proper convergence only at the centre of the screen. Away from the centre the red, green and blue patterns produced by the three guns become more and more displaced from each other and it is necessary to carry out dynamic convergence to superimpose them to give a single white pattern.

This is done by feeding currents of special waveforms at line and field frequencies, and derived from the line and field timebases, to the six coils of the convergence assembly and the blue lateral coil. In all there are 14 controls to be adjusted to secure dynamic convergence and during the process some readjustment of static convergence may be needed. Unfortunately, the settings of many of the controls are interdependent and the newcomer to convergence does not find their adjustment at all easy. The main thing is never to attempt to correct any particular fault completely by one control. Convergence can only be achieved gradually. It is necessary to correct each fault in turn only partially and to keep going over the whole set of controls time after time.

In addition to convergence grey-scale tracking is needed. This is something else which is absent in a black-and-white receiver. The three guns in the tube cannot be identical and adjustments are needed to the relative amplitudes of the video signals applied to their cathodes and to the voltages applied to their first anodes to compensate for their discrepancies. If adjustments are made merely to produce a white raster at full brightness, then instead of dimming off through all shades of grey as the brightness is turned down, the raster will acquire a colour cast. The relative cathode drives and the anode voltages have to be adjusted so that as the raster passes between black and white it does so through all shades of grey.

Before any attempt is made to obtain a colour picture it is essential to obtain a really good black-and-white picture. This may not be quite as good as that given by an ordinary black-and-white receiver; it will almost certainly be less bright and there may be a little colour fringing towards the edges, for perfect convergence is not always obtainable. Apart from this, however, unless the receiver is working perfectly stably and reliably in monochrome, it is a waste of time to try to introduce the chrominance circuits.

We shall not, therefore, discuss the chrominance circuits here, even in outline, and their detailed description will be deferred



A rear view of the receiver without the power unit and the colour circuits. The i.f. and video amplifiers are on the left in the photograph with the timebase unit on the right and the convergence controls above the tube.

until one of the later articles of the series The Wireless World receiver has been built around a 19-inch colour tube. This was chosen in preference to the 25-inch because of its smaller size and weight which made it much more convenient in view of the large amount of handling needed during development. Anyone who wishes to use a larger tube can readily do so, however, for the operating conditions are identical; only two things are different, a larger tube screen and larger degaussing coils are needed. The units containing the circuitry, being made to fit around the smaller tube, can readily be disposed around the larger. One difference in performance may be noted, the larger tube is likely to give a somewhat less bright picture because the same beam power is distributed over a larger area of screen.

A unit construction has been adopted and, apart from the chrominance circuits, the main units are power pack, timebase, convergence, luminance amplifier, i.f. amplifier and tuner. In addition to the deflecting and converging assemblies on the tube neck, there are a few parts mounted on the tube base and degaussing coils are mounted on the tube screen.

It was stated in the announcement in last month's issue that no provision is made for 405-line reception. This has been done chiefly for simplicity, since the switching needed for a dual-standard colour receiver poses severe mechanical difficulties. The real trouble is not so much the amount of switching needed, for it is surprisingly little, but in the many different parts of the equipment where it is needed. The mechanical difficulties arise because the different switches must all be linked together for operation by one control.

Switches are needed between tuners, in the sound and vision i.f. amplifiers, in the video amplifier, in the line timebase and in the convergence circuits, where most controls must be duplicated. The switches tend to occur physically in places where they are very awkward to link together and where it is almost impossible to do so using standard components. Provision for dual-standard operation is, of course, a commercial necessity and it is relatively easy to do under commercial conditions where special parts and linkages can be fabricated.

In any case, it does seem rather wasteful to use a colour tube for monochrome pictures. An expensive tube is having its useful life wasted when a black-and-white picture could be produced at least as well, if not better, by an ordinary tube.

Transistors

So far nothing has been said about transistors. They are in fact used in considerable numbers in the i.f. amplifiers, the early video stages and the chrominance circuits. Valves are used, however, in the timebases and the final luminance and chrominance stages. It is possible to use transistors in all these places and at least one commercial receiver does so. One of the main practical difficulties lies in the fact that a colour tube is subject to occasional internal flash-overs which do no harm to the tube but which can easily kill any transistor connected to it. If transistors are used, therefore, it is necessary to adopt a great many protective devices. At the present time it is simpler to use valves which themselves buffer the earlier transistor stages.

The fact that valves are used does entail one disadvantage. The available valves have heaters designed for series connection and this means that the chassis is necessarily live to the mains. It is hardly practicable to build an isolating transformer into the receiver. The live-chassis receiver is, of course, safe enough when it is enclosed in a cabinet and it is quite normal in commercial practice. If it is not so enclosed, however, and it can hardly be enclosed when fault-finding, it is potentially dangerous. This, of course, is nothing to do with colour. It applies to all mains equipment which does not include an isolating transformer and that is probably the majority of commercial equipment.

However, in the interests of safety the constructor is very strongly advised to use an external 1:1 ratio isolating transformer external to the receiver when working on it. If there are any earthed objects in the vicinity this must be regarded as an essential safety precaution, and it must not be forgotten that an electric soldering iron may very well be an earthed object in the vicinity! If for any reason a transformer cannot be used make quite sure that the chassis is joined to the neutral side of the mains, that there are no earthed objects within reach and that the mains plug of the receiver is completely withdrawn from its socket before using a soldering iron.

Before concluding this first general article, it may be well to give a little more detailed picture of the form of circuitry employed. The tuner itself is a commercial product and it feeds the i.f. amplifier through a short length of coaxial cable in the usual way. The i.f. amplifier has three stages with transistors, using stagger-tuned single circuits as inter-stage couplings and wave traps to provide the necessary out-of-band attenuation.

The vision carrier is placed at 39.5MHz and the sound carrier at 33.5MHz. Those accustomed to 405-line television must remember that standards are different with the 625-line system. Negative modulation is used on the vision carrier which means that the tips of the sync pulses correspond to peak carrier power. Frequency modulation is used on the sound carrier, and this permits the use of what is often called 'intercarrier sound'. With this a common i.f. amplifier is used for vision and sound, but in it the sound carrier is attenuated by about 26dB relative to the vision carrier.

In the vision transmission the amplitude is never allowed to fall below some 5 per cent of its maximum; this occurs during colour bar waveforms. By attenuating the sound carrier by 26dB one ensures that it is always weaker than the vision signal; and, in fact, it must be attenuated more than this.

At the vision detector the vision carrier acts, in effect, as a local oscillator to heterodyne the sound carrier and produce a frequency-modulated sound signal at the difference frequency of 6MHz. This is picked out from the vision detector and passed to a 6-MHz amplifier which has two transistor stages and a ratio detector.

The video signal from the detector is fed to a transistor phase-splitter of about unity gain. The video signal is taken in one phase from the collector to feed the video amplifier proper, and in the other phase from the emitter to feed the chrominance circuits and also a two-stage d.c. amplifier which provides automatic gain control to the i.f. amplifier and the tuner.

The first stage of the video amplifier proper is another transistor phase splitter. Its emitter circuit contains a low-resistance potentiometer for contrast control from which the video signal is taken to the grid of the video output valve. The sync separator is fed from its collector.

Substantially the whole of the video gain is provided by the valve output stage, the valve being of a special high- g_m type. The anode is directly coupled to the tube cathodes, d.c. restoration being effected in the grid circuit.

The bandwidth of the video amplifier approaches 5MHz. The chrominance circuits, which we shall not discuss here, have a much narrower bandwidth. Because of this the time of transit of a signal through the chrominance circuits is about 0.6μ sec greater than through the video circuits. If something were not done about this the colour on the picture would be displaced sideways from the picture detail. What is done is to include in the video amplifier a delay line to equalize the transit times of a signal through the two paths. This is quite a separate and distinct thing from the PAL delay line, which is used in the chrominance circuits for an entirely different purpose.

As stated above, the video or luminance output stage feeds the tube cathodes. This is with the luminance or Y signal. The chrominance circuits conclude with three valve output stages which feed the three grids of the tube with R-Y, G-Y and B-Y signals, so that with Y on the cathodes, the grid-tocathode signals become R, G and B.

When first setting up the equipment, the chrominance circuits are omitted and the three grids of the tube are then returned temporarily to a common fixed potential, so that the tube is fed with the Y signal only at its cathodes. The picture is then monochrome.

The field timebase is fairly conventional and has two valves. One is the output pentode and the other is a triode, the two forming a multivibrator type of sawtooth generator, which is locked by the field sync pulses in the usual way. Outputs are taken from the pentode cathode and from a winding on the output transformer to the convergence circuits.

The line output stage is also basically corventional. It is driven, however, by a LCoscillator running at line frequency which is kept at the correct frequency by the output of a phase detector which compares the line sync pulses with pulses from the line output stage. This is, of course, flywheel sync. While this is usually considered desirable when the signal is subject to interference, it is not usually necessary in areas of high field strength. In this case, however, it is necessary, and for a rather interesting reason.

A timebase which is directly locked by sync pulses is normally set to run at a basically lower frequency so that it can be tripped by the sync pulses line by line before it would trip of its own accord. This means that if the sync pulses fail for any reason, for instance if the transmitter breaks down, the timebase continues in operation at a frequency lower than the proper one. This normally causes no harm, but in this case it results in the anode dissipation of the line output valve becoming excessive. At the lower frequency the interval between successive flybacks is greater and the current has time to build up to a greater value.

Flywheel sync is needed to prevent this, for the free-running frequency with it is very close indeed to the locked frequency.

Space Communication and Travel

Arthur Clarke, who proposed the concept of synchronous communication satellites in his article "Extra-terrestrial Relays" published in Wireless World in October 1945, collaborated with Stanley Kubrick, the film director, to produce "2001-a space Odyssey" which had its gala premiere at the Casino Cinerama Theatre, London, on May 1st. This spectacular space travel film centres around a megalomaniac talking computer which takes over an inter-planetary expedition with disastrous results. The film is both technically interesting and photographically breathtaking as one travels through a galaxy of colours into "space".

Several months ago Mr. Clarke addressed the British Interplanetary Society on "Voices from the Sky—the past, present and future of communication satellites". This paper, which was reproduced in the March issue of the Socity's journal *Spaceflight*, made reference

to his prophetic Wireless World article. Commenting on the article he said "I must confess I . . . thought in terms of manned space stations, because in 1945 there was no alternative in the available technology. It was a good day on our unit [R.A.F.] if only one tube blew in the day's operations. The idea that you could have complex electronic equipment, without hoards of servicemen and spare parts to replace the burnt out components then seemed almost as fantastic as space travel itself. As far as I can remember the immediate reaction to the appearance of the Wireless World article was precisely zero".

The copper plaque reproduced below was recently presented to Mr. Clarke by Dr. Anthony Michaelis, science correspondent of the *Daily Telegraph*. Among the signatures are those of H. F. Smith and F. L. Devereux (respectively, Editor and editorial assistant of *W.W.* in 1945).



Loudness Control for a Stereo System

Circuit for automatic balance and tone compensation at all listening levels

by R. T. Lovelock, C.Eng., F.I.E.E.

It is customary to control the volume of a reproduction system by varying the gain of the audio amplifier while the signal is still at a low level. The standard method of doing this is to apply the signal across the track of a variable resistor, and to tap-off the desired fraction of it between a slider and the low-potential end of the track. In a practical system it is desirable to have an adequate range of controllable variation so that all tastes may be met, and all classes of music catered for. On the other hand it is equally desirable that the range shall not be too great, otherwise a sufficiently precise setting of the level cannot be obtained. A good range for which to design a system is 40 dB.

Very early in the development of electronic reproducing systems it was found that a linear variable resistance was not adequate for the purpose because the ear has a logarithmic response, and the majority of the desired range was crowded into the initial 10% of the rotation. To obviate this trouble the carbon-track resistor with a pseudo-logarithmic law was produced. The required response is approximated by constructing two nearly linear portions of track, of differing slopes, joined at an overlap by a deep curve. Although the response is far from the true logarithmic law and due to variation of the location of the curve joining the two end portions no two resistors have exactly the same law, every resistor gives a

Fig. 1. Typical envelopes for the three types of variable resistor indicated.



Effective rotation (%)

satisfactory spread of voltage levels for the purpose of gaincontrol. One difficult problem in stereo systems is that of ganging two gain controls together. To cater for differences in the overall gain of the two amplifiers a balance control is provided but once having balanced the system at an intermediate output level, it is desirable that it remains balanced at all other usable outputs. This can only occur if the two gain controls ganged together have a very similar law. Unless the system does remain balanced it will be necessary, for good stereo effect, that the balance control be carefully adjusted at each re-adjustment of the output level, and this is an exasperation for the expert, and an insufferable nuisance for all others.

The root of the problem lies in the fact that no two log-law carbon-track resistors have nearly identical laws, and if they are ganged together their outputs will not remain in balance over more than a very limited portion of their usable range. A linear-law resistor would remain much more closely balanced but unfortunately it would not cover an adequate range of gain. The commercial items which the constructor is normally able to purchase are manufactured and supplied to a specification which defines the allowable departures from the theoretical law. Fig. 1 shows the extreme envelopes within which the resistance may fall for log carbon, linear carbon, and linear wire-wound

Fig. 2. Curves showing maximum difference between channels (lower channel as a percentage of higher channel).

100



types. It is very unlikely, of course, that any two resistors will differ by as much as this maximum permitted amount but it is equally unlikely that they will be very close to one another.

When these resistors are used in the conventional circuit for gain control and two of them are ganged together, it is possible to calculate the difference between levels in the two channels which may arise from this cause. Near the two ends there will be little difference, of course, and it is over the central portion of the range that they will differ most. On the other hand, it is this central portion which is most used in practice and which is representative of the performance to be expected. Fig. 2 plots the extreme difference which may be obtained over the central portion of the range, the lower channel being expressed as a percentage of the higher one. Again it is emphasized that it is very unlikely that this maximum difference will ever be experienced but a large difference will be met sufficiently frequently to make it a serious problem. It is evident from this figure that if a linear law is used to avoid the great differences in the log law, then a wire-wound resistor will be very much better than a carbon one.

In an endeavour to gain the advantages of closer matching in wire-wound variable resistors, an attempt was made to develop a circuit which would give a range of gain approximately linear over 40 dB using a single variable resistor in each channel. It was obvious that to attain this end it would be necessary that the resistor controlled the gain by two independent means simultaneously, and this was achieved by causing it to attenuate the signal directly, and also to reduce the gain of the stage by increasing the degree of negative feedback. A schematic of the circuit employed is shown in Fig. 3. The variable resistor has its slider connected to earth thus dividing the track into two separate resistors, each variable in value and connected to earth. One of these portions forms the lower arm of a signal attenuating circuit, while the other forms an impedance in the emitter circuit resulting in negative feedback in the amplifier.

In Fig. 4 the overall variation of gain with rotation of the control is plotted for a nominal law which is linear from zero to $1 k \Omega$. The departure of this law from the ideal linear characteristic, linear over a range of 40 dB, is also plotted. The ideal law is followed so closely that no apparent departure from it is evident when operating the equipment. In Fig. 5 is shown the maximum possible difference of levels between the two channels when a maximum possible difference exists between the two resistors. This is seen to be less than 5 dB for the worst case and is likely to be no more than 2.5 dB for all practical cases. For comparison, the departure between two channels for a log-law carbon resistors. This reveals that the normal circuit using ganged carbon controls can have an error which is 5 times greater than that given by the linear circuit.

It is well known that the sensitivity of the ear does not vary in a linear manner over the whole of the frequency range. The



R. T. Lovelock has been technical director of Belling & Lee Ltd., since 1965. During his career on electronics circuit design he has been involved in work connected with the British Common Standard system since the inception of the second Burghard Committee, and has also been a U.K. delegate to the International Electro-technical Commission Committee meetings on environmental testing for a number of years.



Fig. 3. Schematic diagram of the gain control circuit.



Fig. 4. Curves showing the overall gain / attenuator characteristic.

Fig. 5. Difference of level between channels for maximum spread of variable resistors.



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Fig. 6. Circuit diagram of the loudness control. The $3.3k\Omega$ resistor marked * is replaced in the amplifier by a tone control network with an effective resistance of $3.3k\Omega$ at 1kHz.

difference in levels between the threshold of audibility and that of pain is much less at very low and very high frequencies than it is in the middle of the audio spectrum. If the frequency response is adjusted to sound correct when the reproduction level is high, it will sound thin and attenuated when the level is turned down to a soft effect. Since some people desire a high level, while others cannot endure it, if the response is maintained constant while the level is altered, the reproduction will be correct at only one of the many preferred levels. If quality is to be maintained at all levels it will be necessary to readjust the tone controls for each setting of the gain control. This is a requirement which is as inadmissible for all but skilled users, as would be the need to readjust the balance control, and in the interests of simple operation it is desirable that operation of the gain control should automatically vary the frequency response of the amplifier so that the optimum condition is maintained. Fortunately, the ear is not excessively critical, and it is possible



to design a simple circuit that will be sufficiently approximate to the optimum condition, to be acceptable over the whole of the 40 dB range.

This circuit is shown in Fig. 6. Included is a switch which will allow the automatic correction to be cut-out, leaving the response constant with adjustment of the gain. If this switch is not required, then the additional two resistors and one capacitor associated with it can also be omitted. Two stages of amplification are shown, but the second is not needed if only the gain control is required; in the amplifier from which this circuit was taken, the $3,300\,\Omega$ resistor was replaced by a tone-control network by which means the overall frequency response could be adjusted. As the gain is reduced, both low and high frequencies are reduced less than the central band. The 3μ F capacitor boosts the low frequencies by increasing the impedance of the lower arm of the attenuator with decreasing frequency. The 0.1μ F capacitor boosts the high frequencies by decreasing the impedance of the negative feedback network with increasing frequency. The maximum value of bass-boost is limited by the $1,000\,\Omega$ resistor shunting the capacitor, and the maximum value of treble-boost by the 100 Ω resistor in series with the other capacitor. The undesired bass-boost given at maximum gain is offset by two other corrections; by attenuation due to the 0.47μ F capacitor, and by increase of impedance in the 200μ F feed-back capacitor which limits amplification at frequencies lower than 10 Hz.

The frequency response of the system at the extreme positions of the control, and also in the middle position, is shown in Fig. 7. The full line is the response with the automatic correction operating, and the dotted line is that with it switched out. The fall in gain above 20 kHz is given by the 4.7 nF capacitor, and contributes to the stability of the system. The fall below 20 Hz is of assistance in the removal of transient surges and of rumble; this would be assisted, of course, by the inclusion of a rumble-filter, but it is a wise precaution to back-up such a filter with a falling response in each separate section of the amplifier. If the tone-control of the amplifier is adjusted for acceptable quality at the mid-position of the gain-control, the response will be acceptable to the ear over the whole range of the control. To achieve this end it will be necessary that all inputs to the amplifier have a magnitude which gives maximum amplifier output at maximum gain control setting. In the amplifier from which this circuit was taken, each input channel was fitted with an individual preset attenuator to allow this condition to be obtained.

Wire-wound variable resistors.—Variable resistors normally available through retail channels are the commercial equivalents of models which in slightly different form meet the

> requirements of a military specification. Although they are not specifically sold to this specification, because so many of the parts used are common to both models, the resistance law will be approximately the same. The military specifications currently applicable are DEF 5122 for composition type, and DEF 5121A for wirewound type. The laws and limits assumed in the article are derived from these two specifications, from the preceding RCS series, and from drafts now under discussion in the International Electro-technical Committee and in the British Common Standard. The resistors used were Colvern pattern CLR, chosen for their small size, approximately one inch diameter. The use of a larger type would have given very slightly better performance.

Fig. 7. Gain characteristics of loudness control.

Colour TV Development Continues in Europe

Techniques described at Paris colloquium

Although Europe now has several colour television broadcasting services in operation, and the technology and hardware are beginning to look much the same wherever you go, colour television as a subject for research and development is by no means dead. A few years back there was great technical excitement over the systems battle. That field is certainly pretty well closed, but there is still much to be done in the improvement of electro-optical devices and information processing equipment within the framework of the established systems. Some idea of what fields are being considered worthwhile exploring could be gathered from the Colloque International sur la Télévision en Couleur held in the UNESCO building in Paris from 25th to 29th March. Over three hundred delegates and contributors from all over the world attended this conference, including a sizeable group from the communist-bloc European countries. (Details of Russian colour television sets, presented by Professor S. Novakovsky of the U.S.S.R., were published in the May issue, p. 113).

Display devices. The single-beam colour c.r.t. still holds a good deal of promise as a competitor to the shadow-mask tube, although a great deal of early research seems to have sunk without trace. Philips Research Laboratories, Eindhoven, consider the beam index version a worthwhile project and M. Lubben and P. M. v.d. Avoort described some aspects of their work on this device. As is well known, the beam index tube has a single electron beam scanning a tri-colour screen made up of vertical'stripes of red, green and blue phosphors separated by black 'guard' stripes. In the Philips version all these stripes are 0.2mm wide and there are 400 RGB triplets across the width of a 23-inch tube. When the beam (spot size 0.4mm x 2.4mm) is activating a particular colour phosphor stripe it must be modulated with the colour signal, R, G, or B, appropriate to that stripe, so the tube includes an 'indexing' system to give

information on the position of the beam. This uses an associated stripe pattern on the screen: deposited over every other black guard stripe is an additional phosphor stripe which emits ultra-violet radiation when energized by the beam. This u.v. energy is picked up by a photomultiplier mounted on the outside of the tube envelope, and the output from this—the 'index' signal—indicates the position of the beam and is used for sequentially selecting the R, G, B colour signals.

A block diagram of the selecting system is shown in Fig. 1. The index signal from the photomultiplier is first filtered and limited to remove frequency and amplitude components due to the video modulation of the beam. It is then frequencydivided by 1.5 to make it equal to the triplet scanning frequency needed for switching in each of the colour signals. (The two frequencies cannot initially be made the same, by using one index stripe per colour triplet, because crosstalk would occur and cause phasing problem.) The output of the frequency divider, of angular frequency ω , is then fed into a phase splitter which produces two carrier frequencies displaced in phase, sin ($\omega t - 21^{\circ}$) and cos ($\omega t - 19^{\circ}$). On these are suppressed-carrier-modulated the R - Y and B - Y signals from the colour receiver. The two modulator outputs are then added to produce a chrominance signal, C. This is added to a monochrome signal, M, which is obtained from the luminance signal Y and a colour-difference signal M - Y (derived from matrixing R - Y, and B - Y). In the absence of colour information, say on a monochrome transmission, R - Y, B - Y and C are zero, and M = Y.

The final signal applied as modulation to the grid of the beam index tube is the equivalent of three band-limited amplitudemodulated pulse trains in t.d.m., one giving R samples, one G samples and the third B samples, and it contains a d.c. component, the monochrome information, plus an a.c. component,

Fig. 1. Circuitry necessary for operation of the beam index single-beam colour c.r.t. The photomultiplier provides information on the position of the beam which is used for switching the colour signals.





Fig. 2. Principle of operation of light-valve tube for projection television using the Pockels effect.



Fig. 3. Time-division multiplexing scheme in colour camera channel, allowing R, G and B signals to pass through a common processing unit.

the sequential R, G, B information. The stripe phosphor efficiencies are adjusted so that equal beam current for R, G and B results in a white screen: thus when a monochrome signal is transmitted only the d.c. component is present, and this produces intensity-modulated white; but when colour information is transmitted in addition, the presence of the a.c. component causes this equal-energization balance to be changed so that colour is produced on the screen.

Many of the operations in Fig. 1 are performed digitally, so that, although the circuitry is complex, the number of components required can be reduced by the use of integrated circuits. The main problems with the tube at present seem to be that the vertical stripe pattern is more visible than the dot pattern in a shadow-mask tube (although this can be mitigated by a lenticular screen placed in front of the tube); the saturation of primary colours is slightly lower than the correct value; and there are small errors of hue. But the big fundamental advantages of this single-beam tube—no mask or grid required, lower e.h.t. than the shadow-mask tube for a given brightness, no convergence system needed, no colour or grey-scale tracking problems—must provide a big incentive for continued development of the device.

A tube for projection television, performing a function similar to that of the Eidophor system but using a different physical principle, was described by G. Marie of Laboratoire d'Electronique et de Physique Appliquée (France). It modulates the intensity of light from an external source and can be used either for monochrome picture projection or, in a group of three tubes with an optical colour separation system, for colour television projection. The physical principle used is the Pockels effect, which is similar to the Kerr effect but taking place in a solid instead of a liquid. When certain crystalline materials are subjected to an electric field they cease to be optically isotropic and the refractive index of the material becomes dependent on the direction of propagation and plane of polarization of the light passing through it. The plane of polarization of the transmitted light can be altered by changing the applied electric field and so, by inserting polarizers in the incident and emergent light beams, it becomes possible to modulate the intensity of the light by varying the voltage producing the electric field.

How this principle is applied in the L.E.P. tube is shown in Fig. 2. A beam of polarized light is directed at a monocrystal of KD₂PO₄ mounted inside the evacuated envelope of the tube, and is returned from a reflecting surface on the rear of the crystal, finally passing through a second polarizer. The polarizers are crossed so that when the electric field applied to the crystal is zero no light is transmitted through the system. The video signal voltage is applied between a transparent conducting layer on the front face of the crystal and a grid spaced about 20µm behind the crystal. This structure is scanned by a constant-current electron beam in the normal television manner, and at each point where it impinges on the crystal the beam forms almost a short-circuit between the crystal and the grid. If this point on the crystal is at a potential lower than that of the grid, the crystal loses electrons by secondary emission and its potential tends to rise. Conversely if the crystal potential is higher than that of the grid, the electronics cannot escape and the potential tends to fall. An equilibrium condition is reached, leaving the point on the crystal a few volts more positive than the grid. When the video signal is applied, at each point "touched" by the electron beam the grid-crystal capacitance at that point becomes charged to the voltage of the signal at the corresponding instant. There is consequently an electric field at this point and the Pockels effect occurs. Thus the light transmitted through the system is spatially modulated according to the pattern of electric field "cells" written by the electron beam.

The device is constructed in the form of a 3-inch image orthicon camera tube and uses conventional deflection and focusing coils. Examples of projection pictures it produces were shown. Definition and contrast were described as "satisfactory".

M. T. Inamiya described a Japanese portable colour receiver, intended for the American market, which has been designed around a 12-inch shadow-mask tube developed by Toshiba. This rectangular tube is 35cm long, has a 90° deflection angle and weighs 3.6kg (8lb). The electron guns are smaller than normal (8mm in diameter) but the tube neck diameter is standard—36.5mm. Unipotential focusing is used in the guns. Average anode current is 400μ A. The shadow mask contains 117,000 holes of 0.24mm diameter, with a pitch of 0.65mm. Phosphor dots on the screen are 0.41mm in diameter and have a pitch of 0.67mm. The faceplate glass is tinted, giving 65% light transmission.

Cameras. A miniature Plumbicon pick-up tube with a diameter of 1.6cm-half that of a standard Plumbicon-has been developed by Philips Research Laboratories. At the colloquium its possibilities were strikingly demonstrated by a Philips man who produced a portable colour television camera, using three of the tubes, which was comparable in size and weight with a 16mm film camera without lenses. Despite the small size, the definition is said to be "not much different" from that of a large camera, while sensitivity is "comparable". The camera itself weighs 3kg, while an associated electronics unit connected to it weighs 4kg. Applications in television news broadcasting and closed-circuit systems are envisaged. The miniature tube, which is 13cm long, has an image diagonal of 10.5mm and uses a modified version of the photoconductive layer of the standard Plumbicon. The mesh screen behind the layer-part of the anode -has an extremely fine structure of 1500 meshes per inch. Focusing is electrostatic. Tube definition is given as a modulation depth of 40% at 3MHz with a maximum "white" signal current of 0.2μ A.

An interesting innovation in a four-tube Plumbicon camera recently introduced by C.S.F. (France), and described by R. Cahen, was the use of a single electronic processing unit in the camera channel for all three colour-tube signals. red, green and blue. Normally cameras have a separate processing unit for each of these signals, but where the signals undergo simultaneous non-linear operations (e.g. gamma correction) in these units it is difficult to avoid discrepancies between them and there is consequently a danger of colorimetric distortions occurring. In the C.S.F. camera this problem is averted by multiplexing the separate R, G and B signals into a composite signal which then passes through the single processing unit so that all three colour components receive identical treatment. The combining and eventual separation of the R, G and B signals is done by time division multiplexing, as shown in Fig. 3. Each colour signal is sampled at a p.r.f. of 5MHz, and the three pulse trains, each carrying its colour information by pulse amplitude modulation, are displaced in phase relative to each other and combined so that the resulting composite signal has a p.r.f. of 15MHz. This then passes through the processing unit and is finally de-multiplexed by a reverse operation to recover the separate R, G and B signals.

Home video recording. A technique designed to allow colour television signals to be satisfactorily recorded by and played back from domestic video tape recorders was outlined by M. F. Koubek of the Technical University of Vienna. Present domestic v.t.rs can cope with monochrome signals but cannot handle signals including a colour subcarrier (at, say, 6MHz) because of their restricted bandwidth-typically about 2MHz. Shifting the colour subcarrier for recording purposes down to about 1.5MHz would, of course, result in a very visible dot pattern on the picture. Koubek's solution is to encode the narrow-band colour information-actually three colour difference signals —on a line sequential basis, à la SECAM, to produce a chrominance signal which is combined with a luminance signal at the lower end of the video frequency band. The sum of the three colour-difference signals is zero, so the average value of the chrominance signal is zero, and consequently the composite signal, to be recorded and played back, is compatible. Since the chrominance information is at the lower end of the video band, the composite colour signal is not seriously impaired by bandwidth limitations in the recording equipment.

Fig. 4 shows the principle of the encoding system. The R, G and B input signals, obtained from a receiver, are transformed by matrixing into a Y (luminance) signal, a colour difference signal U=0.19(B - Y) and a colour-difference signal V = 0.51 (R - Y). The third colour difference signal, W = (G - Y), is obtained from the other two, by addition and polarity inversion, in order to ensure that U + V + W = 0. Thus, while hue and

Fig. 4. Encoding scheme for producing a narrow band composite colour signal suitable for home video recorders.





New C.S.F. four-tube Plumbicon colour camera utilizing the multiplexing principle shown in Fig. 3.

saturation are conveyed by U and V, the signal W makes the system compatible. The three colour-difference signals are then selected in turn, on a line sequential basis, and added to the Y information to form the final composite signal. Koubek explained that the frequency spectrum of the chrominance signal does not contain frequency components at multiples of the line scanning frequency, so that the chrominance spectrum can be frequency intercalated with the luminance spectrum. The decoding equipment, for recovering the R, G and B signals from the played-back information, utilizes two delay lines. Apart from domestic recording, Koubek suggested that the technique could also be used for narrow-band industrial television systems.

June conferences and Exhibitions

Further details are obtainable from the addresses in parentheses LONDON

June 5 & 6	Northern Polytechnic
Space Communications (Dept. of Electronic & Communications Engg, Northern P	olytechnic, London N.7)
June 10-14	Savoy Pl.
Industrial Measurement Techniques for on-line Comput (I.E.E., Savoy Pi., London W.C.2)	ers
June 27 & 28 Better Management Through Standards (B.S.I., 2 Park St., London W.1)	Imperial College
OVERSEAS June 3-8 Electromagnetic Wave Theory (Int'l Scientific Radio Union, 7 Pl. Emile Danco, Brussels 18)	Stresa
June 10-14 British Engineering Exhibition (S. Black, London Chamber of Commerce, 69 Cannon St., L	Copenhagen ondon E.C.4)
June 12-14 Communications Conference (I.E.E.E., 345 E. 47th St., New York, N.Y. 10017)	Philadelphia
June 17-19 Microelectronics Symposium (I.E.E.E., 345 E. 47th St., New York, N.Y. 10017)	St. Louis
June 20-22 Optimal Systems Planning (Prof. T. J. Williams, Laboratory for Applied Indust University, Lafayette, Indiana 47907)	Cleveland rial Control, Purdue
June 23-26 Impact of Microelectronics (Electronic Industries Assoc., 2001 Eye St., N.W., Washington	New York n, D.C. 20006)

Wireless World, June 1968

Output Transistor Protection in A.F. Amplifiers

A description of protection methods, outlining their limitations, and a practical circuit providing a linear load-line characteristic

by Arthur R. Bailey,* M.Sc.(Eng.), Ph.D., M.I.E.E.

THE sensitivity of transistors to overload has been one of the major factors restricting their use in power amplifiers. Indeed many amplifiers have warnings stating that damage can result from accidental short-circuit of the output connections. In fact, failure of the output transistors may be due to one of two major causes, and it is important to distinguish between them.

The most obvious cause of failure is due to overheating by excessive collector dissipation. This usually occurs in class B output stages, when the collector dissipation can rise to excessive values if the load is accidentally shortcircuited.

In class A operation, the collector dissipation in a transistor amplifier is the same under quiescent conditions as when driving into a short-circuit load. If severe overdrive occurs, the stages normally back-bias automatically and limit the output current to a safe value.

Unfortunately class A operation necessitates very large heat sinks to absorb the high continuous power dissipation of the output transistors. To reduce this dissipation to more reasonable proportions the so-called pi-mode

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Fig.1. Normal positions for mean overcurrent protection devices.

Fig.2. Peak overcurrent protection using diverting diodes.



method of operation has been used.¹ In this case the output transistors operate in pure class A for small input signals but progresback-bias with increased signal. sively Ultimately full sine-wave drive causes the operating point to be that of pure class B. The quiescent class A current is approximately the same as the mean current under class B conditions so the mean transistor current is approximately constant. This therefore protects the transistors from overload so far as mean dissipation is concerned as overload has little effect on the mean transistor current. Unfortunately the dissipation in the transistors is still quite high and there is also the possibility of the circuit biasing back into class C for full drive on waveforms other than pure sine-wave.

Class B is the obvious choice so far as dissipation is concerned and nearly all commercial amplifiers run in or near to pure class B operation. Unfortunately in class B there is no protection of mean transistor dissipation by the biasing circuits and the system is wideopen to catastrophic failure.

Mean overcurrent protection

The most commonly used protection system is that of circuit disconnection if the mean output current exceeds the correct full-drive value. Such disconnection can be in the powersupply, output transistor circuits or the output lead of the amplifier. The mode of operation may be a fuse, thermal trip, electronic trip or other device. Such protection may be satisfactory if the power transistors have a large reserve of dissipation, but for many transistors the mode of operation is too slow and the output transistors may fail before the protection operates. Even if failure does not take place, the transistor junctions may have been damaged and the device may therefore fail at a later date. Such protection is therefore inherently unsafe.

Peak overcurrent protection

This is less common than the previous type, but it has been used commercially by Grundig and also by the author.² In this mode of protection, diodes are used to prevent the peak output current exceeding a pre-determined safe value. This prevents high peak currents flowing in the output transistors, and materially assists in their protection. Unfortunately the mean output current is then higher than the normal full sine-wave drive value and the transistors may fail due to excessive dissipation if the overload is prolonged. Backing-up fuses are therefore desirable to prevent breakdown on prolonged faults. Basic circuits for mean and peak overcurrent protection are shown in Figs. 1 and 2. It should be noted that there are really two classes of operation under the heading of "mean protection", mean d.c. supply current being true mean protection, whereas fuses are really r.m.s. protection. They have however been grouped together as in neither case is there any attempt to limit the peak transistor current on overloads.

It might be expected that peak current protection with backing-up mean protection would be sufficient for any power transistor. Unfortunately this is not so, and the reason cannot be found in mean collector dissipation. There is, therefore, a second effect tending to cause failure and this is the phenomenon known as secondary breakdown.

Secondary-breakdown in power transistors

In transistors having a considerable junction area mean collector dissipation is no longer a complete guide to safety. It is possible to have different parts of the wafer at differing temperatures forming local "hot-spots". Also at high working voltages the effective base-width of the transistor is reduced by "Early-effect".3 This increases the nonuniformity of current distribution across the transistor wafer; such irregularities occuring naturally due to imperfections in base width and doping levels, etc. The non-uniform current distribution tends to generate local hot-spots and the localized heating increases the current gain and the tendency to avalanche breakdown at these points. The increased localized current gain increases the current density in the hot-spots and thermal runaway may ensue giving failure at the hot-spot. This localized thermal avalanche breakdown can occur at mean power levels far below the power that the transistor can dissipate at low values of collector voltage.

Curves of Safe Operating Area Ratings (SOAR) are therefore given for many output transistors, the worst case for breakdown being at d.c. and low frequencies. This is



Fig.3. Safe operating areas and maximum dissipation line for typical output transistor (2N 3791).



Fig.4. Resistive and reactive load-lines for class B output stage showing breakdown risk with pure reactive load.

because short pulses do not give the hotspots time to establish fully, and the consequence is that failure is more likely in low frequency operation. Audio-frequency amplifiers come within the worst area of operation so it is essential that output transistors are kept within the worst SOAR characteristic. Typical values are shown in Fig. 3. It will be seen that the maximum allowable d.c. dissipation at high collector voltages is far less than the normally quoted "maximum dissipation". Fixed current-limit transistor protection will therefore be unable to protect adequately under reactive load conditions, where large currents may be drawn at the time when the collector voltage is a maximum.

Reactive load conditions

Under resistive load conditions in a class B stage conduction takes place when the transistor collector is above half the rail-to-rail voltage. Secondary breakdown is therefore extremely unlikely. Reactive loads however draw maximum current when the voltage across the transistor is a maximum and this is the worst case for secondary breakdown. Unfortunately all loudspeakers and most laboratory dummy loads are not resistive and therefore an amplifier may test safely on a resistive load but blow-up on a loudspeaker. Anyone doubting the reactive properties of loudspeakers should consult Briggs' book on loudspeakers⁴ where some very alarming degrees of load reactance are shown with normal types of loudspeaker.

Fig. 4 shows the operation load lines for one half of a class B output stage under conditions of equal load impedance; in one case pure resistive and the other pure reactive. The conduction at high collector voltages is very apparent in the reactive load case and could obviously cause failure. For foolproof protection it is therefore essential to limit the operation of the output transistors to within the d.c. safe operating area, irrespective of the load impedance presented to the amplifier.

Load-line protection

Plain overcurrent limiting has shown to be inadequate at high values of collector voltage, so some method of reducing the limiting current value at high voltages must be used for complete protection. Ideally the limit to output current would be along the limiting line of the SOAR characteristic, but unfortunately this is normally non-linear. In practice, however, a straight line will enable sufficient output power to be developed; this limiting load-line is shown in Fig. 5. Provided that a circuit will limit the output transistor current to this load-line, then there is no possibility of transistor failure due to overload.

The simplest way of attaining such protection is to use a transistor as a shunt across the drive to the output transistors. If the base of this transistor is fed with voltage signals proportional to both output transistor voltages and current, then the transistor will start to conduct at a point which follows a linear load line. A suitable circuit⁵ is shown in Fig. 6. The emitter-base conduction voltage of approximately 0.6 volt (for a silicon transistor) determines the point where protection starts. The base of the protection transistor is driven nearer to conduction by either increased collector current or collector voltage of the output transistor. The current sensing is done by utilizing the voltage drop across the emitter resistor of the output transistor, this being proportional to the emitter current (and therefore approximately collector current) of the output transistor. By a suitable choice of component values the protection line can be arranged to be of any desired slope. To prevent the driver stage overdriving into the protection transistor, a current limiting resistor may be required in the feed to the base of the output transistor.

Practical protection circuits

It must be understood that no universal values can be given that will protect all amplifiers, owing to the widely differing voltage and current levels involved. It is possible however to calculate the ends of the limiting load-line as follows:—

$$I_{max} = \frac{0.6 (R_2 + R_3)}{R_2 R_e}$$

$$V_{max} = \frac{0.6 (R_2 + R_3) R_1}{R_2 R_3}$$

Assuming the value for R_e that is given for the amplifier design and using a value of 47 ohms for R_3 , then the value of R_2 can be evaluated knowing the maximum allowable d.c. collector current. Inserting the value of V_{max} then enables the value of R_1 to be determined.

Often the SOAR ratings are not known, and then one has to resort to a certain degree of guesswork. To allow for driving into fairly reactive loads I_{max} must be greater than the peak current output into a resistive load say 50% greater. I_{max} is therefore approximately:

$$I_{max} \approx 2.25 \sqrt{\frac{\text{output power}}{\text{load resistance}}}$$
 amperes

e.g. 2.25 amperes with an amplifier rated for 16 watts output into 16 ohms load. Equally V_{max} can be allowed to be about 20% greater than the rail-to-rail h.t. voltage (e.g. 60 volts for a 50-volt h.t. rail difference). These values appear to be quite safe in practice and no failures have been obtained.

In the circuit design described by the author² this method of protection can be applied instead of the current limiting diodes originally specified. A suitable circuit is shown in Fig. 7. This operates very satisfactorily and is certainly safer than the original protection which has not always protected on gross overload.

Under conditions of full drive into shortcircuit or reactive loads the transistor dissipation will be high. The value will be greater



Fig.5. Use of linear limiting load line to give safe operation.

Fig.6. Linear load-line protection circuit using clamping transistor across input.





Fig.7. Practical protection circuit for transformer-driven amplifier. Thick lines indicate added components.

Fig.8. Load-line protection for quasi-complementary output stage.



Thick lines indicate added protection circuits

than the maximum obtained with sine-wave drive into resistive load, so the heat-sink area must be adequate if continuous drive into short-circuits is envisaged. This dissipation will depend on the values of current and voltage limit used, but will normally be about 12 watts for each output transistor in a nominal 16-watt amplifier. Heat-sink size is therefore dependent on the duration of any possible overload. For normal service there does not appear to be the need to cater for such sustained overloads, but laboratory amplifiers will obviously need this reserve.

All the tests conducted by the author have shown the protection to be absolutely safe, even though sparks can be drawn by shorting the amplifier output! The system will operate up to very high frequencies if planar transistors are used for clamping the drive. Equally very sharp spikes fed to the amplifier are limited safely without any breakthrough. The circuit therefore appears to afford the complete solution to output transistor protection.

For complementary and quasi-complementary output stages it is necessary to use complementary clamping transistors. These are readily obtained at reasonable cost and can be used to protect the driver transistors as well as the power transistors. A suitable circuit arrangement is shown in Fig. 8. This has been used very successfully by the author. In conclusion the author would like to acknowledge the use of the facilities of the University of Bradford where much of the testing was done. Thanks are also due to Mr. Arthur Radford for his encouragement and his many transistors sacrificed in the search for an adequate system of protection.

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3. M. J. Morant "Introduction to Semi-conductor Devices" (Harrap 1964) p.92

4. G. A. Briggs "Loudspeakers" Fifth Edition, (Wharfedale Wireless Works) pp.78-85.

5. S. G. S. Fairchild Ltd., Design Note 16-20 Watt High-Fidelity Amplifier using Silicon Planar Transistors. June 1966.

Interference and Microwave Systems

The congestion of the frequency spectrum presents an ever increasing problem to the microwave transmission engineer because of the risks of interference. Such interference may be within a single system, or between differing systems and, in this respect, interference between terrestrial line-of-sight systems, and between terrestrial and satellite systems warrants particular attention because of the ever increasing need for communication capacity. For example, the current Intelsat satellites receive transmissions radiated by earth stations, in the 5925-6425-MHz band, and re-radiate the the 3700-4200-MHz transmissions in band. However both these bands are allocated on a shared basis to terrestrial services and are already extensively used throughout the world. This means that there is a considerable risk that terrestrial systems, in the 4-GHz band, may (if their aerials are pointing towards the satellite) cause interference to the satellite. Similarly, earth stations radiating several kilowatts of power in the 6-GHz band may well interfere with nearby 6-GHz terrestrial systems. A conference was held in London at the I.E.E., on April 23rd and 24th, to discuss some of these problems.

In many cases it is possible to optimize aerial design in order to reduce interference, and the first session of the conference was devoted to these designs. A paper by Dr. M. S. Afifi (Technological University of Delft) discussed the problems of scattering from microwave aerials, particularly in relation to a design employing a paraboloid and a plane reflector. Although the superior sidelobe performance of such "folded" types has been accepted for some time, the cost would probably be prohibitive when applied to a satellite earth station. G.F.S. Swann and M. Flack (G.P.O.) reported the results of measurements, made with the Goonhilly aerial, on two transmitters at 167 and 230 km distance. In each case the aerial was tracked over a square grid and locii of percentage of time, for which a specified signal level was exceeded, were plotted. There was marked evidence of reflections from a 1000-ft television mast some 1.2° off the direct bearing.

Two papers by A. E. Baker (G.P.O.) discussed the problems of planning terrestrial radio relay systems so as to avoid interference from within, and outside, the system. Potential sources of interference within the system include the effects of local oscillator harmonics, feeder mismatches and image and adjacent channel crosstalk. Again, extensive re-use of the same frequency is aided by good aerial discrimination.

The Intelsat II and III satellites employ the multi-destination carrier principle, i.e. a transmitting station assembles a composite baseband, consisting of channels for a number of destinations; each of the distant stations receives this composite baseband and selects those channels that are destined for it. This arrangement presents a number of frequency planning problems, since it is often difficult to select a frequency which is acceptable to all those who have to receive it (because of the risk of local interference from other terrestrial systems sharing the same band). The procedure adopted for the Intelsat system was described by J. B. Potts and J. F. Arnaud, of COMSAT.

Interference between satellite systems is possible should two satellites lie within the beamwidth of the earth station aerial. This fact will ultimately place a limit on the number of satellites that can be usefully placed in the synchronous orbit, and this subject was discussed in a paper by J. K. S. Jowett, of the G.P.O. It was shown that from 2-4 satellites could be placed in a 10° arc, of the synchronous orbit, and that these could carry some 10,000-20,000 circuits.

A factor of considerable importance to the multichannel telephony systems engineer is the relation between wanted to unwanted r.f. carrier ratio, and the interference level produced in a telephone channel. This is frequently called the "interference reduction transfer factor", and an analysis of the factor for interference to, and from, wideband and narrowband f.m. and s.s.b. systems, was given by P. B. Johns (University of Nottingham).

A session was devoted to the siting of earth stations. In the U.S., for example, where there are some 5,000 and 2,500 terrestrial radio-relay systems already operating in the 4 and 6-GHz bands respectively, co-ordination raises considerable problems. The difficulties of siting military stations, which may have to operate in close proximity to h.f. stations, radars and other navigational aids, were described by representatives of H.Q. Signals Command, R.A.F. A paper by D. E. Cridlan (G.P.O.) described the results of measurements on the spurious outputs of radar systems. Those employing magnetrons can be particularly troublesome as many of the unwanted products bear no harmonic relation to the fundamental frequency.

In general, it was apparent that microwave system planners are fully aware of the potential risks of interference, but one wonders whether some of the system designs have not, perhaps, too many built-in safety factors.

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News of the Month

B.O.A.C. Flight Booking System

It would appear incredible an airline could consider that spending $\pounds 42M$ on an automated flight booking system was an economically viable undertaking. B.O.A.C. obviously do as it is costing them this to have BOADICEA (British Overseas Airways Digital Information Computer for Electronic Automation) installed. This computer complex will link all major B.O.A.C. flight reservation centres in North America and Europe with a central computer terminal at Heathrow Airport London.

Three I.B.M. computers in London will communicate over voice quality telephone lines with Ferranti Argus 400 computers at the main flight booking terminals. Each Argus, in turn, will drive a number of Ferranti c.r.t. display systems and provide the computing capacity necessary for the keyboards associated with these displays and for organizing the information into blocks for transmission to the central computer.

In use, a booking clerk interrogates the system using the keyboards, perhaps in answer to a customers enquiry for a flight between two points on a given day. Within three seconds the c.r.t. will show details of all suitable flights and the number and types of seat available. When the customer selects a particular flight the clerk types in other details together with any special requirements (diet, etc.,) the disc files of the central computer are updated and a flight booking confirmation is immediately sent off to the relevant display equipment. The computer operates in a "conversational mode" that is, it "asks" the clerk for the next piece of information; this greatly helps in any complicated multi-flight journey bookings.

At Heathrow "computer central", three I.B.M. computers are being installed with a comprehensive disc storage system of 1,400 million byte capacity (1 byte = 8 bits). The disc stores will hold details of B.O.A.C. flight schedules for almost a year in advance together with a record of the number of seats booked on each flight and other relevant information.

Information is exchanged between computer central and the Argus satellite computers at the rate of 2,400 bauds, corresponding to 400 characters or about seven messages, a second; it is expected that this will be increased to 4,800 bauds later.

The difference between this and other

flight booking systems that have been built in the past can be summed up in two words, speed and flexibility. The main parameters of the system are set entirely by the software, modifications being performed by a programme alteration. This policy is reflected right through to the keyboard/display units. All that these store in the way of characters are sixteen lines of different length at 52 different angles, therefore any drawing, shape of character can be built up from these lines as a programme on the Argus 400.

The flexibility is illustrated in the method used for training booking clerks. Here an Argus 400, driving a number of displays, has been programmed as a teaching machine. Booking clerks are presented with a series of typical booking situations and are asked to go through the correct procedure. In the event of a mistake being made the clerk is informed of his error by the Argus which will not proceed with the programme until the error has been corrected.

Incidentally, B.O.A.C. intend to extend the use of the system to flight planning, cargo control and other day-to-day problems sometime in the future.



A Ferranti c.r.t. display unit which has been installed as part of the terminal equipment for Boadicea. The equipment consists of some 30 Ferranti Argus 400 computers and 700 c.r.t. display units.

Applying "Bosworth" in Semiconductors

Starting in September in Southampton is the first course to be run by a newly formed training organization called the Semiconductor Technology Centre. Operated jointly by Southampton University and Associated Semiconductor Manufacturers Ltd., this centre is an attempt to put into practice the recommendations of the Bosworth Report* on getting graduate engineers and scientists into industry. The Report said the most pressing problem was to devise some kind of effective "matching" between the output of university and the input of industry, and this is what the new centre hopes to provide. No actual building has been put up for the purpose but the 16-month M.Sc. course will use facilities provided by the University's Department of Electronics and by the Southampton factory of A.S.M. Ltd. (Joint directors of the project are Professor G. D. Sims, of the University, and Dr. E. A. O. Roberts, of A.S.M.) The industrial and university work done by the students will, however, be integrated at all stages: lectures will be given by people from both sides and the whole course will be under joint supervision. The organizers stress that it is not a sandwich course.

Financial support, to meet the capital and running costs of the scheme, is being provided by the Ministries of Technology and Labour, the Department of Education and Science (University Grants Committee and Science Research Council), the Engineering Industry Training Board, and the Conference of the Electronics Industry. Running expenditure will be recouped from the fees charged to students, which will be £1,800 per student (£500 at the University, £1,300 at A.S.M.). Most of the students will be graduates just about to be recruited into the semiconductor industry, and will be sponsored by the various firms who will be employing them. Technically, therefore, the fees will be paid by their employers (in addition to their salaries), but in practice the fees will be largely offset by a number of grants for which most students are expected to qualify (e.g., £500 from the S.R.C. to pay the University fee; a £500 Industrial Studentship grant and a £540 grant from the E.I.T.B.).

*Under the chairmanship of Mr. G. S. Bosworth a Working Group set up to examine the education and training requirements of industry issued two reports: an introductory one, "A review of the scope and problems of scientific and technological manpower policy", H.M.S.O. Oct. 1965, and "Education and training requirements for the electrical and mechanical manufacturing industries", H.M.S.O. 1966.

GAD—The Printed Transistor

Thick film circuits incorporating both passive and active thick film devices have become a distinct possibility. This fact emerged at the well attended joint I.E.R.E., I.E.E., conference on thick film technology held recently at Imperial College, London. In a paper prepared by G. A. Wilkin and R. J. Mytton of the International Research and Development Co. Ltd., and G. H. Elson of Rutherford College of Technology, work being carried out at I.R.D. on thick film transistors was described.

In a similar project, being carried out at Texas Instruments, U.S.A., the acronym GAD (Graphic Active Device) was suggested as being suitable for transistors produced in this manner as the initials t.f.t. have already been reserved for thin film transistors. The aim of the I.R.D. project is to produce an insulated gate field effect transistor with a gain bandwith product of 10MHz when operated in the enhancement mode.

The semiconductor material chosen for producing these devices must be capable of being formed into an organic based ink for screen printing on to a ceramic substrate. The screened layer, after firing, should yield a sintered polycrystalline layer with good semiconductor properties. Although the bulk material may possess such properties it does not follow that they will remain after the printing and firing process. At I.R.D. the substances being used are cadmium sulphide and associated chalcogenide materials.

Three approaches are being made to the problem of forming a suitable dielectric layer. The first of these entails printing a relatively thick dielectric layer using a normal ferro-electric ink. The second method consists of passing a mixture of silane and nitrogen over the semiconductor printed substrate at a temperature of 400°C forming a layer of silicate on the semiconductor surface. The third possibility is to form a layer of cadmium fluoride by treating the printed substrate with hydrofluoric acid.

The source and drain connections will probably be formed from a gold paste, the chief problem here being to form a short source-drain channel without any short circuits.

The material for the gate electrode is not critical the chief requirement being for accurate dimensions and close registration.

With the technology at its present state it would appear that devices having a transconductance of 0.5mA/V and a gain bandwidth

product of 1MHz could be produced, further development being needed to bring all the variables in the process under control.

British Compact Computer Introduced

Dr. Jeremy Bray, Joint Parliamentary Secretary to the Ministry of Technology, opened the new Computer Technology factory on the industrial estate at Hemel Hempstead on May 1st. At the same time Computer Technology's first product, the computer Modular One, made its debut. Computer Technology was formed when a small group of computer engineers got together, under the leadership of I. M. Barron, and formed their own company. The project was supported financially in Britain by Technical Development Backers and on the Continent by European Enterprises Development; in addition there was also a number of private backers. It is surprising when the current trend is towards larger and larger computer firms that such a small organization managed to obtain this necessary backing in their attempt at going it alone. Executives of the Company say that because the computer industry relies on innovation rather than on research and that because a smaller company can have a faster response to new ideas they consider they have a definite advantage.

Modular One was designed for multiaccess operation and to be competitive with American compact computers. With its smallest store, 8k bits, a central processor and a tele-typewriter it costs \pounds 13,000. It is built entirely using Motorola MECL 2 integrated circuits. In order to cut down costs multilayer printed boards have been avoided and double-sided boards with printed through holes have been employed. Also, by careful design the need for interboard back wiring has been reduced to a minimum and is carried out on more double-sided printed boards or by flexible printed wiring that

One of the more complex configurations of Modular One. The desk arrangement houses a processor and three 8k store modules plus a high-speed tape reader. A high-speed tape punch and a second processor are at the rear.



results in reduced production costs and increased reliability.

Modular One has a store cycle time of 750ns, an addition time of 1.5μ s, it will multiply in 2 to 3μ s and divide in 3 to 4μ s. Data is transferred at a rate of 22M bits/sec and an infinite number of programme interrupts can be accommodated. A sixteen bit word is used. Programmes are written in SYMBOL or FORTRAN and run under executive control with communication via the terminal. Any number of independent programmes can operate concurrently and the system changes context in 4μ s on interrupt.

Extra storage units, peripheral and central processors can be added as desired to expand the machine. Each unit is supplied complete with individual power supply and integral interface unit so that expansion may be carried out with the minimum of trouble.

Droitwich Frequency Stability

The signals from the B.B.C.'s 200kHz transmitter at Droitwich have long been used as a reference frequency standard. Since 1965 the long-term stability has been within ± 5 parts in 10^{10} and with the use of automatic frequency correction the excursion from nominal has not exceeded 1 part in 10^{10} .

The frequency stability has again been improved by using a rubidium gas cell standard provided by the National Physical Laboratory. This has a day-to-day stability better than 1 part in 10^{11} and the frequency will be maintained within ± 2 parts in 10^{11} of nominal referred to the caesium beam standard at N.P.L.

The phase of the received signal is monitored at N.P.L. and the value of mean daily frequency is available on application to The Director, National Physical Laboratory, Teddington, Middlesex.

Yachtsman's Electronic Gadgets

One of the contenders in the third singlehanded transatlantic race which starts from Plymouth on June 1 is a 49-year-old electronics engineer, Noel Bevan, of Hartley Wintney, Hants. His boat Myth of Malham is equipped with a number of home-made electronic aids one of the most interesting being an audible warning device, that rings a bell if the boat goes off course. Such a thing is not new but to produce it as cheaply as Bevan has-using neither of the established principles, i.e., concentrating the earth's magnetic flux in a soft iron bar or using a compass controlled device-is noteworthy. He made the device, which he calls an offcourse indicator magnetic field sensor, for under £10 and it includes a unit which dedecides whether the swing off course is merely momentary before waking the skipper.

There is no engine aboard, and Noel Bevan has designed a device for charging his batteries by converting the heat from his pressurized paraffin heating stove to current, based on "back-to-front" connected Frigistors.

There are seven radio receivers in the



Home-made equipment that will assist Noel Bevan in his bid to win the transatlantic yacht race. Top, thermo-electric battery charger. Left-hand column starting at top; directional radar receiver, omni-directional radar receiver with alarm, broadcast receiver, vlf. receiver, df. receiver. Other items shown are; communications receiver, electronic thermometer (water temperature), frequency standard, anemometer, and offcourse indicator.

boat. One gives audible warning if a ship carrying radar comes within 10 miles and another gives a bearing on that ship. The radar warning receiver uses a horn aerial and employs a reflex type circuit; when a signal is received the horn doubles as a "loud-speaker", the audio transducer being mounted in its base.

Another low-frequency receiver is for the reception of naval and other station broadcasts on weather conditions.

For iceberg warning, he has underwater acoustic apparatus similar to the sonar "ping" acoustic set-up in submarine chasers, only looking forward instead of downward. There are two quartz crystal chronometers, one homemade. Wireless World hopes to be able to give constructional details of some of these devices sometime after the race.

New Zealand in Satellite Link

New Zealand will probably decide next month to open a satellite communications link with the world. The Post Office in Wellington has recommended a start on a satellite ground station and ancillary equipment rather than a second Trans-Tasman undersea cable. Grounds for the decision are that there will be a slight economic advantage. Australian telecommunications engineers visited Wellington about two months ago and discussed with the N.Z. Post Office officials, who had earlier despatched two of their own engineers to Britain to study satellite systems, the possibility of a satellite link. Three years ago New Zealand became a member of Intelsat and undertook a commitment to pay about \$A640,000 over four

years into the "club" without any immediate benefit. The move, however, safeguarded national interests in future satellite transmissions, and rights can be exercised at any time. It could be three or four years before satellite transmissions can be commercially exploited in N.Z. The link envisaged will be primarily for burgeoning telephone and telegraph traffic, although it will have a capacity for television, including colour television. So far the N.Z. television system lacks even a network hookup between its four channels. It has to be decided between the Post Office and the Broadcasting Corporation who will be responsible for the necessary local landline or microwave links, which would relay any satellite-borne programmes.

Now its Plug-in to Expertise

We are all very familiar with the concept of building up complex systems with electronic building blocks, plugging in, or wiring in, modules to achieve our aims. The technique has spread into other fields; soon, if the educationalists have their way, we will be building up our knowledge by "reading in" clearly defined small blocks of knowledge and each budding specialist will receive his training from a selection of such standard educational modules. This became clear at a recent I.E.E.T.E. conference on modular education at Loughborough University of Technology.

Discussions at the first session ranged over the problems of the present course structure; was it satisfactory; is the present selection system good enough; how long should an educational module be and whether a single course structure could be provided for the wide spectrum of activity covered by technicians.

Later discussion hinged around the problems related to training engineers for particular branches of industry and as to whether a single formula could be produced, within the modular scheme, to cover all aspects of training, retraining and continuing training.

Two facts that emerged were that modular education had a vital part to play in the future and that educational establishments should concentrate on educating, and industry on training.

Missile Defence Contract

The British Aircraft Corporation have won a £150M contract that entails supplying a mobile and flexible air defence system to the government of Libya. The system consists of Thunderbird and Rapier anti-aircraft missiles controlled by dual-purpose radars which, together with a comprehensive telecommunications system, provide early warning and tactical control for missiles and fighter aircraft. Sub-contractors to B.A.C. are Marconi, Elliott Automation, and Plessey Radar. They will be responsible for the radar, communication and computer controlled display equipments. Ferranti are involved in the production of Thunderbird, and Decca Radar, Barr and Stroud, and Cossor have an interest in Rapier.

The Royal Television Society invites applications for the 1968/69 John Logie Baird Travelling Award. This award, value £200,

is open to post-graduate students (in United Kingdom educational establishments) who are concerned with some scientific aspect or aspects of electronic engineering, television or allied technology. It is expected that the award will be made to someone in the age group of 21-30 years of age. The Award is intended to assist the successful applicant in undertaking a period of investigation abroad of approximately 6-8 weeks during 1968. During this period some aspect of electronic engineering, television or allied technology will be studied or surveyed. Application forms for the Award are available from the Royal Television Society at 166 Shaftesbury Avenue, London W.C.2. All applications must be submitted not later than 31st May.

The possibility of low-priced television receivers from Hong Kong competing in our already over-competitive market has now become a certainty. A company called Promotors Ltd., officially opened last November in Hong Kong, is now producing dual standard monochrome receivers with 9, 19 and 23 inch screens under the brand name Peacock with a target output of 1,500 receivers per month. The Company intends to exploit the estimated £1.5M Hong Kong home market and to export to other Asian countries before launching its assault on the American and U.K. markets. At the present time, in Hong Kong, the small nine-inch portable sells for £38 as against £58 for an equivalent Japanese model and the 23-inch dual standard set costs £68 compared with $\pounds 82$ for the Japanese equivalent.

"Electronics in the 1970s" is the title of an I.O.R.E. convention to be held at Cambridge University between July 2nd and 5th. The convention will consist of a number of whole and half-day symposia devoted to: computers and automation, electronic engineering education, automatic test equipment, communications and future materials and components. In addition, survey papers will be presented on lasers, underwater acoustics, high-power ultrasonics and oceanography. Registration forms are obtainable from the Conference Registrar, 8/9 Bedford Square, London W.C.1.

Users of Mullard educational films should note that a new distributor has been appointed-Educational Systems Ltd, ESC House, Imperial Drive, North Harrow, Middlesex, and that several new titles are available. These are: Principles of Magnetism, Colour Television and Semi-conductor Physics-a five part film. Educational establishments undertaking computer projects will be interested to know that Mullard are offering ferrite core stacks with a capacity of 4k words of 41 bits at reduced prices. The stores are surplus to requirements and are of obsolescent design, interested parties should contact Mr. N. F. Thompson, Mullard Ltd, New Road, Mitcham, Surrey.

In the article a **Sensitive F.E.T. Volt**meter, by D. E. O'N. Waddington, published last month type TIS67 transistors were specified for Tr_1 and Tr_2 . This was in error, the type that should be employed is TIS68. Also, note that the wiper of $S1_b$ should be connected to Tr_3 collector.

Personalities

Douglas C. Birkinshaw, M.B.E., M.A.(Hons.), F.I.E.E., who is retiring from the B.B.C. on June 7th, has been closely associated with British television since the B.B.C. launched the experimental transmissions using the Baird 30-line system from a studio in Broadcasting House in August 1932 when he was in charge of the project. In preparation for the start of a regular television service, Mr. Birkinshaw joined the team responsible for planning the Alexandra Palace station. He was appointed engineer-in-charge in March 1936, and supervised the start of the service in November of that year. During the war he was engineer-incharge at Daventry. In 1945 Mr. Birkinshaw returned to Alexandra Palace as superintendent engineer, television, to supervise the engineering arrangements for the reopening of the television service in June 1946. He relinquished his activities in television in 1963 on his appointment as general assistant to the Director of Engineering.

G. T. Waters, B.E., M.I.E.E., is appointed director of engineering of Radio Telefis Eireann, the Irish broadcasting organization. After graduating from University College, Dublin, in mechanical and electrical engineering, he spent a short time in the Department of Posts and Telegraphs before joining Radio Eireann as transmitters engineer in 1957. In 1961 he transferred to the television

G. T. Waters



service and in 1965 became manager of the Television Production Planning and Control Department. He has been head of production facilities since last August.

J. M. Westhead, B.A., Ph.D., until recently general manager of the Distribution Equipment Division of A.E.I. Cables Ltd., has joined Pye Telecommunications Ltd. as general manager with a seat on the board. Dr. Westhead studied at Oxford University and in 1950 graduated with first class honours in physics. Two years later, while a University demonstrator in electronics, he gained his doctorate in nuclear physics. In 1956 he joined B.T.H. which became part of A.E.I. C. A. W. Harmer, O.B.E., who has been acting managing director of Pye Telecommunications (now part of the Philips organization) has retired. Mr. Harmer has been a director of several of the Pye companies for some years.

H. J. Jones, B.Sc., production manager of the Plessey Components Group's Semiconductor Division at Swindon for the past five years, has been appointed works manager of the division and is now responsible for its overall administrative functioning, as well as for production. Mr. Jones joined the company 23 years ago at its Allen Clark Research Centre at Caswell. A chemistry graduate of Hull and London Universities. Mr. Jones, who is 38, is also a Fellow of the Chemical Society, and an Associate of the Royal Institute of Chemistry.

Eric Tyler, M.I.E.R.E., who began his industrial career with Decca in 1947 after demobilization from the Royal Navy and joined Plessey in 1965, has been appointed group commercial and marketing executive of Plessey Electronics Group. He became commercial manager of the Decca Navigator Company in 1948 and was at one time executive vice-president of Decca Radar in the United States. From 1961 to 1965 he was manager of the New Products Division of Decca Radar as well as international sales manager of the company.

Jack Dinsdale, B.A., M.I.E.E., well known to readers of Wireless

World for his amplifier designs, has joined the staff at the College of Aeronautics, Cranfield. He is a senior research engineer with the Industrial Unit in Precision Engineering and will be concerned with electronics, automation and data processing in relation to precision mechanical engineering. After graduating at Trinity College, Cambridge, in 1959 Mr. Dinsdale joined the Weapons Division of Elliott Brothers where he was latterly project leader in the Military Data Systems Division. Soon after joining Elliotts he did a year's postgraduate study on magnetic tape recording at the College of Aeronautics.

D. E. Todd, B.Sc., M.I.E.E., has succeeded **J. Redmond** as B.B.C. assistant director of engineering. (Mr. Redmond became director of engineering on the retirement of Sir Francis McLean.) Mr. Todd, who is 52, obtained his degree at London University and joined the B.B.C. in 1946 after eleven years on the development of high-power transmitters with Standard Telephones & Cables Ltd. Since joining the B.B.C., he has been concerned with the planning and installation of the Corporation's network of television and



D. E. Todd radio transmitters, and since 1965 has been head of the Transmitter Planning & Installation Department.

W. Peter Dean, who went to the U.S.A. in 1959, has joined the Philco-Ford Corporation's Microelectronics Division as manager of the hybrid circuit department. Since being in America he has been with several semiconductor manufacturers and was, until joining Philco-Ford, product manager of hybrid integrated circuits with the National Semiconductor Corporation.

Waldo Thorn, M.I.E.E., has been appointed managing director of Celdis Ltd., Reading, Berks, in succession to Clifford G. Bailey, who is to devote himself fully to the activities of the parent company, Unitech Ltd. Mr. Thorn, aged 36, became general manager of International Rectifier Company (G.B.) in 1963 and a year later manager of International Rectifier Corporation



W. Thorn

Italiana SpA., Turin, Italy. He left International Rectifier in 1967 and returned to the U.K., and prior to joining Celdis, was responsible for setting up a new micro-joining equipment division in the K. & N. Electronics Group.

George W. Tillett, who contributes our bi-monthly "Letter from America", has been appointed executive vice-president of Audio Dynamics Corporation, New Milford, Connecticut. He joins Audio Dynamics from the Fisher Radio Corporation where he was director of engineering at their Pennsylvania plant. Prior to going to the U.S.A. he had been technical director of Wharfedale, and also chief engineer of Heathkit in Great Britain.

Arthur C. Edwards has retired from his post as commercial director of Eddystone Radio Ltd. A wellknown amateur transmitter (G6XT), Mr. Edwards, who is 62, has been with Eddystone since 1927. He has been commercial director since his return from service in the Fleet Air Arm during the second World War. Eddystone has been a subsidiary of the Marconi Company since 1965.

A. D. Horn, who joined the Radar Division of the Marconi Company in 1952, has been appointed commercial manager. In 1954 he transferred to the Export Department and spent four years in Iraq as a representative of the Company. Since 1963 he has been manager of the Commercial Sales Division. Mr. Horn, who is 44, served in the Radar Branch of the Royal Navy after which he spent five years with B.O.A.C. and International Aeradio before joining Marconi.

Cable and Wireless Ltd., have announced the secondment of **E. P. Eades,** F.I.E.E., to the International System Management Division of COMSAT in Washington, D.C. Mr. Eades, who is 52, has been with Cable and Wireless since 1933. He has served at many overseas stations, was appointed manager in Singapore in 1964 and since April 1965 has been special representative, Far East, based in Hong Kong.

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Marconi—**space communications systems** The Marconi Company Limited, Space Communications Division, Chelmsford, Essex AN 'ENGLISH ELECTRIC' COMPANY

WW----099 FOR FURTHER DETAILS

How Pye record

Pye Records Limited make all their recordings today, whether in the studio or, as in the photograph, on location, exclusively on 'Scotch' Dynarange magnetic tape. It is one of the most important means by which Pye can give the public the high quality they expect from discs bearing the Pye name.

Dynarange has a unique low-noise oxide that gives a signal-to-noise ratio 3 to 5 db better than any conventional tape. This reduction in background noise means improved frequency response, particularly in the higher frequency range, and considerably increased dynamic range. At the same time the exclusive 'Super-life' coating reduces magnetic oxide 'rub-off' to a negligible amount, so the life of the tape is extended, while recording heads last much longer and require far less frequent cleaning.

All these features make 'Scotch' Dynarange magnetic tape today's most advanced recording tape, and the first choice of the majority of recording studios. If Dynarange tape can be of assistance to you, please post the coupon below for technical literature and 'Scotch' magnetic tape data sheets.



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Test Your Knowledge

Series devised by L. Ibbotson*

1. Transistor theory: bipolar and field-effect

1. A bipolar (junction) transistor when used as an amplifier has:

(a) both junctions forward biased

(b) both junctions reverse biased (c) the emitter-base junction forward biased and the base-collector junction

reverse biased (d) the emitter-base junction reverse biased and the base-collector junction forward biased.

2. In a bipolar transistor the amount of doping in the base is much less than in the emitter. This is:

- (a) so that the base will have a high resistance
- (b) to give the transistor a high breakdown voltage
- (c) so that the current crossing the emitter junction will consist mainly of carriers injected into the base

(d) so that there will be plenty of minority carriers in the base.

3. To obtain efficient transistor action the base width must be small. This is:

(a) to keep the overall resistance of the device low

(b) to keep the overall dimensions of the device small

(c) so that most of the injected carriers will reach the collector without recombining

(d) to keep the input and output capacitances small.

4. Three of the following together determine the value of the current amplification factor of a transistor (α). Select the "odd man out".

(a) The fraction of the emitter current carried by carriers entering the base from the emitter.

(b) The fraction of injected carriers which reach the collector without recombining.

(c) The equilibrium density of minority carriers in the base.

(d) The number of carriers produced in the collector junction by the avalanche effect (impact ionization).

5. If the reverse bias voltage between the

collector and base of a transistor is increased in magnitude, the effective base width

- (a) is unaffected
- (b) increases
- (c) decreases

(d) in some transistors increases, in others decreases.

6. In a drift transistor the base doping is graded, the number of impurity atoms per unit volume being greatest at the emitter junction and least at the collector junction. The purpose of this is:

- (a) to give a low input capacitance
- (b) to produce a "built-in" electric field
- which speeds the carriers through the base
- (c) to make the maximum voltage which can be applied to the collector high

(d) to give a low bulk resistance to the base with a high injection efficiency.

Silicon planar transistors are made by diffusing impurities into the silicon. As a result of this method of construction, all silicon planar transistors

(a) must be n-p-n

- (b) are drift transistors
- (c) have uniform base doping

(d) are symmetrical (collector and emitter roles can be reversed).

8. In a p-n-p transistor carrying a steady emitter current the hole concentration in the base is increased above its equilibrium value. As a result of this, the electron concentration in the base

- (a) increases
- (b) remains constant
- (c) decreases
- (d) drops to zero.

9. In a junction field effect transistor in normal use the junction between gate and channel

- (a) is unbiased
- (b) is forward biased
- (c) is reverse biased
- (d) may be forward or reverse biased.

10. A junction f.e.t. has the gate correctly biased and a current flowing in the channel. The channel width

(a) is uniform

(b) is least near the source

(c) is least half way between source and drain

(d) is least near the drain.

11. In a junction f.e.t. "pinch-off" occurs when:

(a) the gate-channel junction breaks down (b) the number of carriers in the channel drops to zero

(c) the depletion layer extends completely across the channel at one point

(d) the depletion layer fills the whole channel.

12. In an insulated-gate f.e.t. (an m.o.s.t.) which is of the n-channel enhancement type the bulk material is:

- (a) n-type silicon
- (b) p-type silicon
- (c) intrinsic silicon
- (d) silicon oxide.

13. In an enhancement insulated-gate f.e.t. the "turn-on voltage" (or "offset voltage") is:

(a) the least drain voltage which must be applied before a current will flow

(b) the least gate voltage required to form a channel

(c) the gate voltage required to produce "pinch off"

(d) the gate voltage required to break down the gate insulation.

14. A depletion insulated-gate f.e.t. differs from an enhancement insulated-gate f.e.t. in that:

(a) in the depletion type a channel exists when the gate bias voltage is zero

(b) in the depletion type pinch-off does not occur

(c) in the depletion type a current can flow between gate and channel

(d) the enhancement type can only be made with an n-channel, the depletion type with a p-channel.

15. Insulated-gate f.e.ts are very easily destroyed when being handled, and when being connected into a circuit. This is because:

(a) they are mechanically very frail

(b) they are easily damaged by heat

(c) the wires are joined to the device in such a way that they break off very easily (d) a small charge accumulating on the gate electrode can produce an electric field large enough to destroy the insulating layer.

16. The main advantage of f.e.ts over bipolar transistors is that

(a) they have a better high frequency performance

(b) they have a very high input impedance (c) they are less sensitive to temperature

changes

(d) they are more reliable.

^{*}Mr. Ibbotson is a senior lecturer in electrical engineering at West Ham College of Technology, London E.15.

Letters to the Editor

The Editor does not necessarily endorse opinions expressed by his correspondents

Audio Myths, Maths & Measurements

With respect to your editorial in the April issue, the following observations may be of interest. A worthwhile subjective improvement was made to a Tobey-Dinsdale amplifier by the use of selected special OC28s with an $h_{\rm FE} \approx 90$ in place of the original AD149s. Tests were then made to ascertain if any instrumental measurement could correlate with this improvement, and it was found that the response was maintained to 70kHz within trace thickness with the selected transistors and to only 50kHz with the originals. Square-wave performance at 25kHz was similarly improved.

It is of interest to note that the preamplifier had the customary 20kHz roll off and that the improvement between the two amplifiers was confirmed by listeners whose hearing cut off as low as 12kHz and also with programme material cutting off at 5kHz.

Miss Barbara Hayes (of C.B.S. records) suggested that the perfect amplifier would have infinite bandwidth and hence any improvement giving increased bandwidth was a step towards the ideal.

It may only be concluded that a bandwidth far beyond the audio passband is of subjective value in improving the naturalness of the sound of musical instruments, and the clearness with which they may be heard. IVOR ABELSON,

Fidelity Radio Ltd.,

London, W.11.

As Mr. Abelson refers to the Tobey-Dinsdale amplifier we asked Mr. Dinsdale for his comments. He replied:—

I was interested to read Mr. Abelson's letter, which in many ways confirms my own observations.

First, I have always preferred the 'sound' of OC29s in my own power amplifier, as giving a subjective audible improvement over the more commonly used OC35s and AD140s. I have also been impressed with the RCA drift-field transistors type 2N2147, the excellent high-frequency response of which gives a noticeable 'cleanness of sound' to my design, although a few minor alterations to certain component values are necessary for optimum results with these devices.

I would therefore agree that a bandwidth

beyond the audio passband appears to give more natural sound to music. However I believe that this is mainly due to the fact that the higher open loop gain at the upper audio frequencies makes for less distortion. Although the power content of music at frequencies above 5kHz is only a few per cent of the total, the human ear seems to be very sensitive to harmonic distortion at these high frequencies. This applies especially to crossover distortion, which incidentally the use of high-frequency output transistors helps to minimize.

With regard to amplifiers of infinite bandwidth, I agree that in the *ideal* situation amplifiers should have an infinite bandwidth and no controls whatsoever (not even volume). However, in our imperfect world, where electro-mechanical transducers have resonances at the upper end of the spectrum, and where motors rumble at the lower end, some form of limit is desirable. Furthermore, when imperfections in the original programme material, in the recording itself, and in the playback conditions are considered, some method of curtailing the frequency response is not only desirable but essential. It is an interesting thought that if the 'ideal' system response extended to d.c., then the displacement of the pickup element as it is lowered onto the groove would be accompanied by a corresponding linear displacement of the loudspeaker speech coil away from its equilibrium position.

No, Sir, by all means let us have wider bandwidths in the power amplifier to minimize high-frequency distortion, but let us maintain in the pre-amplifier (where distortion is negligible) a realistic control over the frequency bandwidth to compensate for the imperfect conditions in which we live and listen.

JACK DINSDALE

Farnborough, Hants.

'Doctoring' Recorded Sound

The item under this heading in the March issue made absorbing reading for all technically interested record users. I recognize that in a short article it is impossible to deal with all the technical tricks adopted by professional recording engineers—even if one could penetrate the secrecy barrier still prevalent in many studios—but I think the omission of any reference to the important Dolby A.301 Signal/Noise "Stretching" System cannot go unmentioned, as the EMT "NoisEx" system is referred to.

Dr. Ray Dolby's audio noise reduction system is now employed extensively in the Decca studios, London, and by many European and U.S.A. recording, broadcasting and film companies for master recording, dubbing, transfer to discs, etc. This elegant system-with its overall 10 dB reduction of noise, hiss, etc., in the mid-range, rising to 15 dB at 15kHz-results in a much cleaner, more transparent recording, with unaltered frequency response and signal dynamics, as listeners to many recent Decca operatic and orchestral LPs will testify. Details of the technique have appeared in the BKSTS Journal, Audio Engineering Society Journal, and elsewhere *

Skilful "doctoring" has been very successfully employed recently by the big recording companies in transferring historic 78 r.p.m. discs to composite LPs, and I have found that practical suggestions for the best pickups, styli sizes, filtering circuits, have a tremendous fascination today for collectors of old-style discs. But the major companies are loath to publish any technical information on methods!

DONALD ALDOUS Audio Record Review, London, W.C.2.

*See W.W. Dec. 1966, p. 632.--ED.

Dry-joint Locator

Now that printed circuits are so commonly used in all fields of electronics, the 'dry joint' fault is even more prevalent than before. This is mainly due to the fact that printed circuits are usually soldered by the solder-bath technique, and unless the print, the component leads, and the solder temperature and cleanness are perfect, invisible dry joints occur.

Field service technicians often find it easier to locate dry joints by checking the continuity of joints and print after the approximate location of the fault has been determined. This is bearing in mind that, apart from transistors, components have a very low defect rate nowadays.

It was therefore necessary to build a continuity measuring device with the following specification:—

1) Low voltage on the test points (to protect in-circuit transistors)

- 2) High current (to persuade possible dry joints)
- 3) No set zero
- 4) Portable
- 5) Inexpensive

In the accompanying circuit a simple voltage regulator, a resistive bridge, and a 100-0-100 μ A meter are used. The whole circuit draws app. 250mA from a 1.5V cell. In the prototype, the bridge supply remains stable with the cell down to 0.9V.

The bridge balances with 1Ω on the test points A & B. With a short circuit on A & B, R_1 (app. 0.5 Ω) was selected to give + f.s.d. on the meter. With A & B open circuit, due to a slight rise in the bridge supply voltage, a



shunt of app. 4Ω was fitted between A & B to obtain a -f.s.d. The voltage and current at A & B are app. 0.2V and 100mA.

The exact values of all the components are not marked on this circuit because they depend on the transistor and Zener diode used.

This circuit has shown many would-be dry joints by indicating fluctuations up to 1 or 2Ω thus indicating them before they give trouble.

D. GOODMAN,

Tel-Aviv, Israel.

How important is detection?

Dr. R. C. V. Macario compares the diode detector with other systems and is critical of the performance of the diode. His diode detectors in Figs. 2 and 3 do, of course, fail to follow the modulation envelope at high modulation depths and high modulation frequencies. This, as he says, is due to the choice of the capacitor across the load resistor. This capacitor is not essential. Without it the detector output is average instead of peak, resulting in a slight loss of audio output but completely eliminating the envelope following problem.

Dr. Macario gives a formula for the optimum value of this capacitor. It shows that the detector time constant should be about 7μ sec for an audio frequency of 1kHz. His detector has a time constant of 75µsec in Fig. 3 and 27μ sec in Fig. 2. Why so large when 7μ sec is about optimum?

The mathematics is difficult to follow because the symbols are not clearly explained.* This is not important because the mathematics is not relevant when comparing the homodyne and the linear diode. Both are zero crossing detectors and the mathematics of one applies exactly to the other. The only practical difference which does not show in the mathematics is that the diode needs several volts to make it linear and will distort on small signals whereas the homodyne will distort on large signals, because the diodes in the shunt detector will be driven into conduction by the signal. Thus the maximum linear signal handling will be determined by the size of the square wave driving the diode bridge as stated at the top of page 56, not by R_1 and R_2 as stated in the caption to Fig. 8.

The comment (page 56) that a.g.c. is made more accurate is not clear. There is approximately a 20:1 loss of signal due to the 4.7k Ω and 2.2k Ω resistors and the 600 Ω filter. Combined with half-wave rectification this gives an a.g.c. of about 2.5mV for the nominal carrier voltage of 100mV and, of course, less for a smaller signal. This is far too small to work the a.g.c. directly and so a high-gain, high-stability d.c. amplifier will be needed in the a.g.c. loop, a further complication. This small signal is partly a result of the use of a shunt modulator. A larger output could be obtained from a two transformer balanced modulator.

The main criticism of the article is that the analysis and waveforms in Fig. 7 show that the behaviour of the homodyne circuit of Fig. 8 is exactly the same as a half-wave rectifier and that it should suffer from distortion during selective fading exactly as would any other envelope detector, e.g. a linear diode detector. To test this I constructed the circuit of Fig. 8 and tested it during simulated selective fading. To simulate fading, a very narrow rejection filter was placed between the signal generator and the detector. The filter was slowly tuned through the signal causing phase shift and then rejection of each component in turn. The second oscilloscope beam was used to show the envelope or diode detector output.

As was expected the homodyne output was exactly the same as the diode output. Both suffered the same severe distortion, both showed modulation frequency doubling when the carrier dropped out.

Can Dr. Macario explain why his homodyne circuit gave "an extra sharpness and greater degree of intelligibility" and why he claims that there is "more faithful following of the signal during carrier fade" when analysis and my experiments indicate that there is no difference?

M. D. SAMAIN

University of Salford, Lancs.

I feel that Dr. Macario may be interested in my own analysis of the operation of the homodyne demodulator. This, I think shows more exactly the similarity between the homodyne and the envelope detector, and leaves me with some doubt as to the credence to be placed on the results of the listening tests.

The basis of my analysis is that the zero crossing signal derived in the homodyne is the carrier on which the received envelope information can be considered to be modulated. The zero crossing signal can therefore be represented as $\frac{e_a(t)}{E(t)}$ where $e_a(t)$ is the received signal and E(t) is the resultant of envelope detection. The output of the homodyne, after the product detector, is thus given by $\frac{e_a(t)^2}{E(t)}$ which can be evaluated using the product form for $e_a(t)$:

$$e_{a}(t) = a \cos(\omega ct + \phi c)$$

$$1 + \frac{m_{u}}{2} \cos(\omega mt + \phi u - \phi c) + \frac{m_{i}}{2}$$

$$\cos(\omega mt - \phi i + \phi c) - a \sin(\omega ct + \phi c)$$

$$\frac{m_{u}}{2} \sin(\omega mt + \phi u - \phi c) - \frac{m_{i}}{2}$$

$$\sin(\omega mt - \phi i + \phi c)$$

This is seen to be E(t), as given by Dr. Macario, in the section of his article dealing with envelope detection, and there are no additional terms which might be attributable

to the term Ψ which was invoked in his analysis of the homodyne's action.

My own view is that there may have been some other mechanism involved in the listening tests. This would seem to involve an oscillatory condition in the zero crossing detector, caused by feedback of the out-of-balance component in the diode bridge circuit. Isolation of this component from the input to the zero crossing detector is dependent on the source impedance, and in the arrangement shown in Fig. 8 of the article would be about 40dB. Assuming, in the listening tests, a direct connection to an i.f. transformer, it can be seen that this isolation is drastically reduced, and in this instance a tuned circuit is also included. This reduction of the stability margin need not be sufficient for a completely self maintaining oscillatory circuit to be established; as indicated by G. Wareham in his comments in the May issue, this does not seem to be a necessary condition for the operation of the circuit to be modified.

It would seem, therefore, that further evidence is required to support Dr. Macario's claims for the operation of the homodyne under selective fading conditions. This evidence should be objective, and it should be supported by a more convincing theoretical treatment than that originally presented. Without this, I, for one, will remain unconvinced of the homodyne's value in broadcast reception, except where the highest possible standards of reproduction are considered to be essential.

E. A. HARMAN

Bolton, Lancs.

The author replies:

ω

A good description of the operation of amplitude detectors is given by W. R. Bennett in the book, 'Communication Systems and Techniques' (McGraw-Hill 1966). In the envelope detector in order that the diode can build up the proper reverse bias on the capacitance C the source impedance should be low whilst the load impedance R_L should be such that C is a short circuit to r.f. signals, but open circuit to a.f. Since the choice of Ccan at best only be pitched between the conditions

$$\frac{1}{\omega_c C} << R_L \qquad \frac{1}{\omega_m C} >> R_L$$

it seemed reasonable to suggest the reactance of C is equal to R_L at the geometric mean frequency $\sqrt{\omega_m \omega_{\gamma}}$ leading to the formula given in the article. It does appear however that some discrepancy has arisen over the values shown in the circuits.

On the other hand the difficulty with the straightforward diode rectifier, as Mr. Samain admits, is the volt or two to make it appear linear as well as the low source impedance just mentioned. In contrast the zero-crossing circuit as described works from a high input impedance and a low voltage swing. (The dotted circuits in my Fig. 8 were for measuring purposes only and one would normally allow much of the 4.7-k Ω resistor to appear as source impedance.) As a result it becomes much easier to organize a.g.c. in the previous stages and keep them linear since the voltage swing is low, and the impedance is high.

The question of overload of the diodes in

^{*}We apologize for lack of clarity in the printing in some parts of this article, especially with subscripts.-ED.

the actual circuit was mentioned in the article. Why the linearity of the detector depends on R_1 and R_2 is because at large signal inputs the μ L 900 amplifier limits one-sidedly, whilst at low signal levels the μ A 710 switches late in one direction. In both instances the zero-crossing signal mark/space ratio changes and so the diodes are not on (or off) for as long. Since R_1 and R_2 fix the circuit gain, the two limits mentioned can be moved up or down accordingly.

Regarding the main contention concerning the similarity of the envelope, linear and homodyne detector, it is well known that the linear diode and homodyne detector are exactly equivalent¹ (at reasonable signalto-noise ratios), because, as mentioned, the signal reinforcement makes the diode(s) appear linear. It is not true to say, however, that the envelope detector is equivalent to a linear detector when transmission impairment exists. The reason is that the linear detector takes note of both the instantaneous signal frequency and the magnitude (envelope). The instantaneous signal frequency, when phase distortion exists, is phase modulated by the term ψ (t). This term is discussed in Professor Cherry's book² which I referred to, and whose symbolism I adhered to closely.

The actual carrier deviation due to $\psi(t)$ is quite small though, and is difficult to observe on an oscilloscope. Using, for example, the figure of $\pm 25^{\circ}$ given in Fig. 5 of the article, one can work out the frequency deviation using an analogous f.m. signal equation,

 $e(t) = \cos (\omega_c t + \beta \cos \omega_m t)$ where $\beta = \triangle f/f_m$. Using the value of 25° makes $\beta = 0.43$, and with $f_m = 1$ kHz, gives $\triangle f = 430$ Hz, i.e. 0.1% deviation. What is perhaps more interesting about the small deviation of 430Hz, at 470kHz, is that to phase detect this information the discriminator requires as demanding performance as does a carrier selection filter, and so was not discussed further at the time. Nevertheless, as shown in Fig. 1, a difference between



Fig. 1. Relative output for envelope and zerocrossing detector for a 50% modulated a.m. signal when the phase of the upper sideband is advanced by 90° relative to the carrier and the lower sideband.

 $E_{v}(t)$ (envelope) and $E_{s}(t)$ (zero-crossing). according to calculation, for the case when one sideband is rotated 90° relative to the other two components, exists. Similar differences also occur in other cases of distortion. Referring to the example given in Fig. 1, however, shows that the apparent depth of modulation (to the listener), which should be 50%, is only 35% in the case of E_{v} , but is 45% in the case of E_{s} . Such analysis only applies, of course, provided overmodulation does not occur. The equations given in the article do not take into account the discontinuity if the phase of the signal reverses, as it can at the onset of overmodulation. The main interest was phase distortion which is something synchronous reception is unable to correct. In fact, calculation of the synchrodyne response E_s corresponding to Fig. 1 almost coincides with E_v and $m_{appareni} = 30\%$.

I feel it is the concern with overmodulation which has caused much of the difficulty. This may well be due to my unfortunate use of the word 'carrier' instead of 'signal' in the last sentence of the article. As far as the detector is concerned, the all important thing is that the 'carrier' continues to switch the diodes. I should add that when the article was submitted, it included a section headed 'conclusions', but this was subsequently used by the Editor as an introduction; if this 'summary' is read at the end, then some of the difficulties are eased.

I am entirely sympathetic to Mr. Harman's last paragraph. I have spent some time looking for reports with objective evaluations of the homodyne system, but it is perhaps worth drawing attention to a review paper on broadcast receivers³ where a number of receiver systems were compared. To quote, '... all receivers... had approximately the same audio response range. The only noticeable difference was that the quality of the homodyne receiver was superior to that of the others'.

R. C. V. MACARIO

1. 'Telecommunications', J. Brown and E. V. Glazier, Chapman and Hall, 1964.

 'Pulses and Transients in Communication Circuits', C. Cherry, Chapman and Hall, 1949.
 'Broadcast Receivers: A Review', N. M. Rust, O. E. Keall, J. F. Ramsey and K. R. Sturley, *J.I.E.E.*, 88 Pt III, 1941, p. 59.

Component Supplies

Your reader's letter in the April issue (p.83) regarding the supply of experimental parts highlights an attitude of mind prevalant in this country and is typified by my recent experience.

I recently purchased an expensive piece of domestic electronic equipment containing transistors from a number of manufacturers. In order to trace a replacement supply of transistors, I wrote to each in turn requesting information and prices.

Of the three firms contacted to date the British transistor firm, a subsidiary of the one from whom I purchased the equipment, ignored my letter. The firm controlled from the Continent sent all the information but ignored a request for prices, probably assuming that I would eventually find a retail supplier. The firm controlled from the States sent me the prices, information and subsequently the spares requested.

J. F. CANNELL,

Mudeford, Hants.

Shortage of Technologists

I was very surprised on reading your February editorial that there is a shortage of scientists and technologists in the electronics industry.

In view of the fact that I have in the past year, applied for six posts with electronics companies and have not been offered a position, I am led to believe that there must be a surfeit of applicants. Now, since it appears that my services will not be required by electronics firms, I have decided to embark upon an alternative career as a lecturer.

I wonder how many other well-qualified applicants receive this sort of treatment and decide to give it up as a bad job. D. G. PAGE Sheffield.

Portable Frequency Standard

As a final comment regarding Mr. Nelson-Jones' "Frequency Standard" may I make the following two points:

(a) Limiting is not a particularly useful method for the removal of amplitude modulation as earlier shown; a quartz resonator should be used in such a way as to allow its intrinsic stability to dominate the application; a "frequency standard" is normally regarded as a monotonic source with a spectral bandwidth less than that obtained by, say, an *LC* oscillator (capable of 1 p.p.m. stability over a period of one second with sound engineering design).

(b) A single quartz resonator at a nominal resonant frequency of 200kHz and with a (typical) Q of 20,000 has a bandwidth of \pm 5Hz, and hence is effective when used to eliminate modulation sidebands of the nature found on the Droitwich transmissions; this use takes *advantage* of the intrinsic stability of the component. The signal obtained in this way, when used as the controlling source in a phase-locked loop containing an oscillator (which need not be one which is stabilized by a quartz resonator) would provide an output signal with a stability at least two orders of magnitude better than the output from the Nelson-Jones device.

The result, for the same price, is a much better performance: This is the essence of my complaint.

Concluding, it should be said that at least three manufacturers, in fact, use the techniques I have outlined to provide an "off-theair frequency standard"; I, therefore, cannot lay claim to any particular originality in my analyses.

LEWIS E. SCHNURR

Mid-Essex Technical College, Chelmsford.

London Meetings

May 22nd. I.E.R.E.—"An adaptive self-phasing array and some applications to radio communication systems" by R. H. Apperley, Dr. D. E. N. Davies and M. J. Withers at 18.00 at 9 Bedford Sq., W.C.1.

May 29th. I.E.R.E. & R.T.S.—"The E.M.I. colour television camera" at 18.00 at London School of Hygiene and Tropical Medicine, Keppel St., W.C.1.

May 29th. I.E.R.E.—"Identifying system dynamics by test signals" by N. Ream at 18.00 at 9 Bedford Sq., W.C.1.

June 14th. I.P.P.S.—Symposium on the various physical aspects of cavitation at 14.15, the Physics Dept., Imperial College, S.W.7.

Decade Frequency Standard

Essential information on a low-cost instrument providing square wave outputs at 100kHz, 10kHz, 1kHz, 100Hz and 10Hz

by Knod-V Weisberg

When making measurements on a.f. amplifiers and the like it is extremely useful to have at hand some reliable means of calibrating oscillators, providing oscilloscope marker pips etc. The instrument to be described fulfils this need at relatively low cost with the minimum of components and provides a useful source of square waves for amplifier testing as a bonus.

The generator consists of a 100kHz crystal oscillator synchronizing four emitter coupled multivibrators each adjusted to divide by a factor of ten. An output amplifier can be connected to any frequency divider or the crystal oscillator by means of a switch thereby enabling any frequency available in the unit to be selected for use. The performance of the instrument is shown in the specification table.

Circuit details

The crystal oscillator (Fig. 1) consists of an emitter-coupled multivibrator with the crystal as the timing element. This has the advantage over some other forms of crystal oscillator in that the positive feedback loop is isolated from the load. The bandwidth of the feedback loop is limited by the 220pF collector decoupling capacitor.

The four frequency divider multivibrators employed (Fig. 2) are all of conventional design and each employs the same circuit, only differing in the values of C_1 and C_2 . The natural frequency of the multivibrators is some 5% lower than that of the synchronizing pulses fed to them, the half period ratios in the region of 47:53%. The synchronizing input signal is arranged to be some 20% of the emitter swing. The output amplifier (Fig. 3) is of conventional design and provides output pulses almost equal to the supply rail voltage.

Specification

		_
output frequency rise time	100kHz, 10kHz, 1kHz, 100Hz, 10Hz ≈0,1µs	
output voltages	mV and 25mV	
power supply	–9, 0, –9V ¹ 240mW	
permissible power		
supply variation	30%	
Internal standard	100kHz crystal	

source before adjusting each multivibrator

in turn, commencing with the 10kHz stage.

The setting of the resistors should not differ

much from a nominal $3.3k\Omega$ to ensure that

the transistors operate on the correct work-

ing point. If it is found that this is not the

case, the value of C_2 should be altered

(normally it will require increasing).

Construction

The prototype generator was built on two 4 \times 6 inch printed boards, the power supply and oscillator being on one of these and the divider chain and output amplifier on the other. The layout of the components is non-critical; however, the emitter connection of the output transistor should be grounded at the output connector and not to the circuit board because of the high current in this transistor.

Point A on the first multivibrator is connected to the output of the crystal oscillator and to the 100kHz position of S_1 ; point B on each multivibrator is connected to point A on the next multivibrator in the chain and also to S_1 .

Calibration

This entails setting each multivibrator to the correct division ratio by adjusting the variable $4.7k \Omega$ resistor in each multivibrator while monitoring the output on an oscilloscope. The first step is to check the crystal output against a known accurate frequency

Fig. 1. The power supply and crystal oscillator circuit. Point A on the drawing is connected to the input of the first frequency divider.



Wireless World, June 1968

0-9V 100kHz from oscillator 820 10kHz B To S CI26 ≥18k ≤i5k ≤33k iokHz IKHZ 100HZ ioHz In In 10n | 100n lOn 100n 1μ ΙΟμ

Fig. 2. Only one of the multivibrator circuits is shown here. The other three are identical with the exception of the capacitance values; these are given in the table.

Fig. 3. The output amplifier circuit.



Hybrid Cascode Amplification

High voltage gain from transistor-valve circuit: Application to power supply stabilization

by B. Alvsten and L. Bergsten



Fig. 2. Hybrid cascode stage.



Symbols used in Text

Ah gml	Amplification of the cascode stage. Mutual conductance of valve V_i
g _{mt}	Mutual conductance of transistor Tr_1
g _{mh}	Mutual conductance of V_1 combined with Tr_1 .
hie, hre	Hybrid parameters of Tr.
hje, hoe	Hyond parameters of 171
Ic	Collector current of Tr_1 .
ib	Base signal current of Tr_1 .
ic	Collector signal current of Tr_1 .
R	Anode load of V_2 .
R'	Bypass resistance.
ro	Collector impedance of Tr_1 .
r in	Cathode impedance of V_2 correspond-
	ing to emitter impedance of Tr_2 .
ra	Anode impedance of V_1 .
v_{g}	Input signal.
Zk	Output impedance of V_1 as cathode follower.

Among the advantages of a cascode stage is its high gain. A conventional double-triode cascode stage, modified according to Attree¹, is shown in Fig. 1. The approximate gain of the stage is often stated to be proportional to the mutual conductance of the lower valve and the load R, but this only holds within a certain range of circuit components as will be shown. (For a list of symbols used in the text see table.)

With the values of components given in Fig. 1, the anode current of V_1 is 2-3 mA, its mutual conductance 1 mÅ/V, and its internal resistance r_a is about 50 k Ω . The input impedance at the cathode end of V_{22} which is a function of the valve parameters and the load resistance R, is here about 12 k Ω . A rough calculation shows that $\frac{3}{4}$ of the signal current from V_1 is usefully working on R, the rest being lost through r_a and R'. The amplification of the cascode stage is about 360 as compared to 470 under the idealized assumption that it should be equal to $g_{ml} \times R$. If anode current of V_1 is raised to 10 mA by a reduction of R' to 10 k Ω , the mutual conductance of the valve rises to 5 mA, and r_a is lowered to about 10 k Ω . Now less than $\frac{1}{3}$ of the signal current usefully enters through R, and in spite of the much larger mutual conductance, the amplification will be only doubled to about 700. To get a higher figure it is necessary to raise the value of R. But this increases the input impedance of V_2 , which means that still more of the signal current is lost in r_a and R'. By circuit adjustments like these, one can achieve an amplification of about 1,500 in practical cases, but not much more. One negative effect of an increased value of R will be a smaller bandwidth. On the other hand, by the addition of a single transistor, which replaces value V_1 , the amplification is easily made one order of magnitude greater, while still keeping Rrelatively small, and by the addition of one further transistor, added to give a "triplecascode" stage, an amplification of 100,000 times becomes within reach. This is possible because the mutual conductance of a transistor is easily made say 50 times greater than that of a valve.

The new stage is shown in Fig. 2, with the transistor fed from a cathode follower. The measured mutual conductance of a transistor BC107 as a function of the collector current is shown in Fig. 3, the upper curve, and in the same diagram the smaller mutual conductance of the combination transistor cathode-follower is also shown. Current through the cathode-follower is 1 mA. A collector current of 2 mA for the transistor, with $R'=100 \text{ k}\Omega$, gives a combined mutual conductance of 50 mA/V resulting in a correspondingly large collector signal current in the transistor. But, as we have already seen, the impedance relationships in the stage are important too. Therefore, the output impedance of the transistor is also plotted as a function of collector current in the same figure, Fig. 3. It is about 45 k Ω at 2 mA, and is a somewhat smaller figure than the corresponding figure



Fig. 3. Showing transistor output impedance r_o and mutual conductance g_{mr} , also the mutual conductance of a transistor and valve combination g_{mh} as functions of collector current.

Fig. 4. The two lower curves show the amplification of the hybrid stage V_1 , Tr_1 and V_2 with two values of R, and the upper curve the amplification of the hybrid stage with the impedance changer Tr_1 included.



Wireless World, June 1968



Fig. 5. Circuit diagram of the amplification stage with Tr, as an impedance changer, cathode follower omitted.

for $\vec{V_1}$ of Fig. 1. The hybrid cascode stage as shown in Fig. 2 has a voltage amplification of 12,000, i.e. one order of magnitude greater than the valve cascode. There is practically no change in frequency response, since this depends in the first place on the anode circuit of V_2 . Let us now look at the possibilities of still higher amplification by adjustments or refinements of the hybrid stage. The important parameters are still the load resistance R and the mutual conductance of the lower active device, here Tr_1 and V_1 taken together.

The two lower traces in Fig. 4 show the change in voltage amplification by the variation of collector current for two values of R: 330 k Ω and 680 K Ω . Nothing is gained by a collector current greater than 4 mA, and the reason for this is the reduction of the transistor output impedance and the external resistance R', whereby an increasing amount of signal current passes through these resistances and is lost as useful current through R.

If a second transistor is inserted as an impedance changer, Tr_2 of Fig. 5, the high mutual conductance of Tr_1 at higher collector currents can, however, be used to advantage in spite of the accompanying low output impedance. Tr_2 has a sufficiently low input impedance to give roughly short-circuit conditions to Tr_1 , and it has, since it is driven by a current source on the emitter side, a high output impedance, of the order of 1 M Ω . Totally, only a small proportion of the signal current is now lost, so that the voltage amplification is again to a good approximation equal to R multiplied by the mutual conductance g_{mh} of Fig. 3. This has been confirmed by measurements, the results of which are given as the top curve of Fig. 4. The curve agrees in growth fairly well with the curve for g_{mh} of Fig. 3. The amplification was measured for $R = 680 \text{ k} \Omega$.

These results have been applied to a small power supply. The power supply was originally built with an ordinary cascode stage. Relevant parts of the circuit are shown in Fig. 6. The modified power supply is shown in detail in Fig. 7. It is seen that the valve V_1 of Fig. 6 has been replaced by

the transistor Tr_1 , and it has been changed to a cathode-follower preceding the transistor. Thereby the voltage-setting potentiometer remains unloaded. It should be emphasized that in the voltage stabilizer of Fig. 7 the high amplification of the hybrid cascode stage is not utilized to full advantage. There is still a considerable percentage change in output voltage for changes in supply voltage, and also for changes in load resistance, and for the same reason in both cases, namely a change in the grid-to-cathode potential of V_1 on account of changing filament voltage, i.e. cathode temperature. A long-tailed pair arrangement would give better results, but probably the best way would be to replace V_1 by a solid-state device, for instance an f.e.t. transistor, adjusted for minimum temperature drift. However,

Fig. 6 (Right). Circuit diagram of the original power supply.

Fig. 7 (Below).

cascode stage.





with or without these refinements the result is a very low output resistance of the power supply, and this has been achieved with very little work and at low cost.

Acknowledgement is due to Professor S. Berglund, who suggested this investigation.

REFERENCES

1. "A Cascade Amplifier Degenerative Stabilizer", by V. H. Attree, *Electronic Engineering*, April 1955, page 174.

APPENDIX

The mutual conductance of a combination cathodefollower and emitter-grounded transistor.

If h parameters are used for the transistor, the equivalent circuit for a cathode follower driving a transistor in common emitter coupling



Fig. 8. Equivalent circuit of a cathode follower and transistor.

Fig. 9. Constant current generator with internal resistance r_0 loaded by bypass resistance R' and input impedance r_{in} of following device.



becomes that of Fig. 8, where the voltage source v_g and the impedance z_k form the Thévenin equivalent of the cathode follower. Note that v_g also is the input voltage on the grid of the valve. In the derivation of the mutual conductance g_{mh} of the combination, the load resistance is regarded as zero, which means that the influence of h_{re} is omitted.

$$i_c = \frac{h_{fe} v_g}{z_k + h_{ie}}$$

and from the definition of mutual conductance we now have for the hybrid combination

$$g_{mh} = \frac{i_c}{v_c} = \frac{h_{fe}}{z_k + h_{ie}} \qquad \dots \qquad \dots \qquad (3)$$

The variation of h_{fe} with collector current is small in the current interval in which the transistor is assumed to operate. Therefore h_{fe} may be treated as a constant. The parameter h_{ie} is roughly proportional to $1/I_c$. With these approximations we get

where a is a constant. This is in good agreement with the curve g_{mh} of Fig. 3.

The voltage amplification of a hybrid cascode stage.

We are now ready to derive an expression for the amplification of our cascode stage, when the lower triode section of Fig. 1 is replaced by a combination according to the above.

Assuming that no grid current is drawn in V_2 , all the current that is fed in to this valve at the cathode, also goes through the anode load resistance R of that valve.

The same is, to a very close approximation, true about the emitter input current of the second transistor, inserted as shown in Fig. 5, if this transistor has a high current gain.

The main interest is now in the collector circuit of transistor Tr_1 , the equivalent circuit of which is shown in Fig. 9. The current generator is v_g , g_{mh} and the input impedance at the cathode of V_2 at the emitter of Tr_2 , in both cases denoted by r_{in} , is seen to be in parallel with the transistor output impedance r_0 , and the bypass resistance R'.

If the parallel combination of r_0 and R' is called R_{t_0} , it is easily found by the law of current division, that

$$i_L = \frac{v_g}{g} g_{mh} \frac{R_p}{R_p + r_{in}} \qquad \dots \qquad \dots \qquad \dots \qquad \dots \qquad \dots \qquad \dots \qquad (5)$$

where $R_p = \frac{r_o R'}{r_o + R'}$

Dividing by v_g we get a transfer conductance for the whole combination $V_1 - Tr_1$, to the input of V, which is denoted by $g_{m\alpha}$

Thus
$$g_{mo} = g_{mh} \frac{R_p}{R_{mb}}$$

The voltage amplification A_h is now simply given

. (6)

Within certain limits $r_{oC} 1/I_x$ and also $R' \propto 1/I_c$. In view of this we get an approximate relationship between A_h and I_c .

 $A_{h} \propto \frac{1}{I_c + b + c/I_c} b$ and c constants (9)

Eq. (9) states that A_h goes towards zero when $I_c \rightarrow 0$ as well as when $I_c \rightarrow \infty$. This indicates that A_h has a maximum, which is in agreement with the measurements presented in Fig. 4.

Both authors are at the University of Lund, Sweden. Bengt Erik Alvesten graduated in physics in 1960 and in electronics in 1964 and is now an assistant in electronics research and education. Lasse O. Bergsten graduated in physics in 1965 and is now doing post graduate research.

Announcements

The 6th International Power Sources Symposium, organized by the Joint Services Electrical Power Sources Committee, will be held at the Hotel Metropole, Brighton, Sussex, from 24th to 26th September. Papers presented will cover a wide range of subjects concerned with research, development and application of primary and secondary battery systems, fuel cells, solar cells etc. Further information and registration forms from J.S.E.P.S.C., P.O. Box 136, Croydon, CR9 2EG, Surrey.

Signal Processing, with emphasis on underwater acoustics, is the subject to be covered at the Advanced Study Institute being sponsored by N.A.T.O., the Raytheon Company of America and the Dutch National Defence Research Organization, to be held in Holland from 12th to 23rd August. The programme includes 35 papers presented by scientists from the U.S.A., Canada and Europe. Application forms and information bulletins can be obtained from the Organizing Committee, c/o Holland Organizing Centre, 16 Lange Voorhout, The Hague, Netherlands.

A vacation school on "Statistical methods of system identification", arranged by the professional group on control theory of the I.E.E. Control and Automation Division, is to be held at the University of Warwick from 16th to 20th September. Further details and application forms may be obtained from the Divisional Secretary, (Control and Automation), I.E.E., Savoy Place, London W.C.2.

A residential vacation school on **electrical measurement practice** is to be held at the University of Manchester Institute of Science and Technology from 14th to 26th July. Further details can be obtained from The Secretary, I.E.E., Savoy Place, London, W.C.2.

The British Amateur Television Club will be holding its 1968 Amateur Television Convention on September 14th in the I.T.A. Conference Suite, 70 Brompton Road, London, S.W.3. Australia's 12th national radio and electronics engineering convention and exhibition will be held at the Wentworth Hotel, Sydney, in May 1969.

At the 23rd Annual General Meeting of the British Radio Equipment Manufacturers' Association on May 2nd the following Member firms were appointed to the Council: A. J. Balcombe, General Electric Co., Hacker Radio, Philips Electronic and Associated Industrics, Pye of Cambridge, Radio Rentaset Products (Baird Television Division), Rank-Bush-Murphy, Standard Telephones & Cables, and Thorn Electrical Industries. Lord Thornycroft, chairman of Pye of Cambridge, was elected president of the Association.

The Muirhead Group, of Beckenham, Kent, have won an order worth approximately £100,000 from the Soviet Union for their latest type of newspaper page **facsimile equipment** and some wire photo products. Publications will be transmitted on this equipment by radio from Moscow to towns hundreds of miles away.

Aircraft navigational beacon transmitters worth more than \$A50,000 have been delivered to the Australian Department of Civil Aviation by Commonwealth Electronics Pty Ltd. The transmitters are non-directional beacons designed to operate in the frequency range of 200 to 415kHz.

A range of tantalum electrolytic capacitors, manufactured by Ero-Tantal GmbH, of West Germany, are now available from G. A. Stanley Palmer Ltd., Island Farm Avenue, West Molesey Trading Estate, Surrey.

Mullard Ltd is to undertake the manufacture of cathode-ray tubes on behalf of Pye; both companies are subsidiaries of Philips Industries. Mullard have acquired the major part of the assets of Cathodeon Electronic Ltd who have ceased to manufacture c.r.ts.

Plans have been made to merge W. G. Pye & Company Ltd and Unicam Instruments Ltd, as part of a policy of rationalization by Pye Holdings Ltd. The new company will be known as **Pye Unicam Limited**.

Tokyo Shibaura Electric, of Japan, manufacturers of **Toshiba** products, have opened an office at Premier House, 12 Hatton Garden, London E.C.1.

Racal Communications Ltd, of Western Road, Bracknell, Berks, have been awarded a contract worth $\pounds 100,000$ by the British Army for the military version of a solid-state h.f. receiver.

Aircraft Supplies Ltd, of 506 Wallisdown Road, Bournemouth, Hants, have taken over the production, repair, maintenance and information play-back of the **Midas** CMM3/RB accident data recording system.

Teleng Ltd are exhibiting a range of v.h.f. communal aerial equipment at a two-day show to be held on August 28th and 29th at the Kensington Close Hotel, London W.8.

Racal Instruments Ltd have signed an agreement to market in the U.K. the programmable voltage sources, known as "Codavolt", and programmable frequency synthesizers, known as "Codasyn", manufactured by the French company Adret Electronique.

Antex Ltd, of Croydon, Surrey, have moved to Mayflower House, Armada Way, Plymouth, Devon.

The Kenya Ministry of Tourism and Wildlife have placed a $f_{30,000}$ order for s.s.b. radiotelephones with the Communications Division of Redifon Ltd. These will link field offices in park and lake areas with the Ministry's headquarters in Nairobi.

The M-O Valve Company Ltd, a subsidiary of G.E.C., have announced the appointment of Metropolitan Supply Company, of 468 Park Ave. South, N.Y. 10016, as **American distributors** for their range of products.

Tectonic Industrial Printers Ltd, of Cirtec Works, Oxford Road, Wokingham, Berks, have changed the company name to **Tectonic (Electronics) Ltd.** The company are mainly concerned with the manufacture of printed circuits.

The electronic research laboratories of Louvain University, Belgium, are organizing a Summer course on "Logic circuits and systems?". Commencing 2nd September, the course will be held in Dutch and French. Further details from Prof. H. P. Debruyn, Electronic Research Laboratories, Section: Circuits and Systems, 94 Kardinaal Mercierlaan, Heverlee, Belgium.

General and technical trends

The possession of good quality sound reproducing equipment is no longer restricted to an exclusive band of enthusiasts with plenty of time, money and that kind of technical expertise which comes from relentless pursuit of an ideal. Anybody who can afford to put down about $\pounds 100$ on a counter can immediately own a quite respectable stereo reproducing chain, as a recent issue of Which? made clear (although a somewhat larger sum of money-about £200-is needed for real hi-fi). This spreading of the benefits of good reproduction to a larger public is, of course, a good thing. It is reflected in the changing character of the annual London Audio Festival and Fair, which from being a small event intended for specialists and run by an engineering society (the old British Sound Recording Association) has now become a large commercial affair open to the general public. (Total attendance over the four days of the 1968 Fair is estimated as 40,000 people.)

This situation is in general deplored by engineers and audiophiles, who tend to see the work they have fostered with love and enthusiasm over the years being gobbled up by big business interests which are concerned only with its profit potential. Of course, this attitude can be criticized as unrealistic. Money must be obtained from selling commercial products to support future technical development. Probably the most important thing is to maintain a good balance between the purely technical and the purely commercial interests. Fortunately, there are still a few audio manufacturing firms in existence which have achieved this balance. The engineers who run them seem to have discovered the secret of being commercially successful without losing their original enthusiasm for technical exploration. Consequently the audio industry still has an engine driving it forward: it has not yet reached the stage of merely free-wheeling downhill under its own dead weight.

As far as the Audio Fair itself is concerned it seems possible that organizational changes are about to take place. C. Rex-Hassan, the director, has hinted that it may be necessary to link up with some established trade—though he has not said what sort of trade he has in mind. He has also complained of the considerable cost of hiring a large hotel such as the Hotel Russell for the event. If this means that the principle of providing rooms for live demonstrations of equipment is to be abandoned, it will be disastrous: without demonstrations the Audio Fair will lose its whole point and degenerate into just another exhibition.

The 1968 Festival and Fair did not reveal any startling technical developments. Equipment is evolving on fairly predictable lines, with continuing emphasis on domestic convenience. The bookcase-size loudspeaker, for example, is now well established, and considerable design effort has gone into making it produce acceptable sounds. Here, stereophony helps considerably because the effect of spaciousness seems somehow to make the lack of bass less important. Nevertheless some manufacturers were brave enough to put on comparative demonstrations of bookcase and larger loudspeakers, and no doubt this could be regarded as a respectable form of "switch selling" for the larger models.

The electronics part of the hi-fi chain can be divided into two main categories of hardware. First there are the systems which use "separates", that is, individual tuners, control units and amplifiers which can be selected separately and assembled into a chain according to the user's requirements. At one time this was the only basis on which the audiophile could build up his equipment, and one which required a good deal of technical knowledge in addition to a desire for high-quality reproduction.

Latterly, from about the time when transistor circuits began to be incorporated in equipment designs, there has been a growing trend towards a second category which combines the tuning, control and amplifier functions in one "integrated" unit, in some cases even including the record player, leaving the user only the task of selecting a suitable loudspeaker or pair of loudspeakers. Both categories are among the new items described on the following pages, which in all but one instance use semiconductors exclusively. Incidentally, some f.m. tuner units favour the use of f.e.ts for r.f. amplifier and mixer stages because these allow large signal handling capacity without introducing crosstalk distortion.

A very large category of equipment at the Fair was tape recorders. Drive belts are at last falling in popularity and we are seeing more three-motor tape transport systems, some with outside rotor motors in all three positions. The resulting reduction in the number of mechanical components leads to quieter machines—a fact that is not revealed

in specification sheets as mechanical noise level is still not given in manufacturers' literature. One company has gone a step further and eliminated the wheels needed for speed change, an all-electronic system being employed. Standardization of equalization characteristics seems to be very much a thing of the future, C.C.I.R., DIN, I.E.C. and N.A.B. standards all being used in the machines seen at the Fair. In at least one case a machine could be switched between two of the standards and in another variable equalization was used. It was disappointing to find no sign of the integrated circuit in tape recorders. However, a few manufacturers are using f.e.ts in low-level amplifier circuits.

Some of the selected items dealt with in the following pages were seen at small individual exhibitions held in other London hotels at the same time as the Audio Fair.

Electronic Speed Control Tape Recorder

The motor speed control system employed in the latest tape recorder from Revox, the 77A, uses no fewer than nine transistors and eight diodes. Full details on how the circuit operates have not been released, but it would appear that the system consists essentially of a closed loop servo controlling the speed of an outside rotor capstan motor. This motor has a number of grooves, or slots, machined in its periphery that rotate in close proximity to a "tachometer type" magnetic head. Pulses appearing at the output of this head have a frequency directly proportional to capstan motor and therefore tape speed. The pulses are applied to a high gain amplifier with a complementary symmetry output stage which is stabilized with a hefty amount of negative feedback. The output of the amplifier drives a frequency discriminator with a d.c. output determined by the difference between the frequency of the signal from the magnetic head and that of the discriminator tuned circuit. This difference, or error, voltage is again amplified and applied to a transistor whose collector and emitter terminals are connected to the + and "outputs" of a rectifier full-wave bridge in much the same way as an s.c.r. is sometimes connected in control circuits. The capstan motor is connected in series with the a.c. terminals of the bridge and the a.c. supply line. The bridge effectively connects the motor in series with the output transistor which then controls the amplitude of the voltage applied to the motor. Speed changes are accomplished by electronically switching, using diodes, a capacitor

into the discriminator circuit which alters the resonant frequency of the tuning element. Two speeds are available, 9.5 and 19 cm/sec that are maintained within $\pm 0.2\%$ by the control system; wow and flutter (weighted) is 0.1%.

A variety of two- and four-track models are available that can be supplied with or without power amplifiers and with a choice of cabinets. The power amplifiers provide 8W per channel (r.m.s.) into $4-16\Omega$ at less than 1% distortion. Equalization is switchable and may be to N.A.B. or I.E.C. standards, the frequency response at 9.5 cm/sec being from 30Hz to 20kHz + 2dB/-3dB. Signal-to-noise ratio, weighted (filter C.C.I.F.), is better than 58dB at 19cm/sec.

Three inputs per channel are provided these being, microphone (switchable Hi $-50\Omega - 150\mu V$ and Lo-100k Ω -2mV), radio (33k Ω 2mV) and auxilliary (1M Ω 40mV). In addition to the loudspeaker outputs, available if the power amplifiers are fitted, two other outputs are provided— 600Ω 2.5V and 2.5k Ω 1.2V

The recorder will accommodate up to 27cm diameter tape spools and is handled in this country by C. E. Hammond & Co. Ltd., 90 High Street, Eton, Windsor, Berks.

W.W. 341 for further details

Richardson Valve Amplifier

One manufacturer who believes that the potentialities of valves have not been fully exploited and that valve circuit designs should not be too hastily discarded is Richardson Electronics. This company, a newcomer to the domestic hi-fi market, was demonstrating a stereo system based on their MA135 valve power amplifier, the circuit diagram of which is shown. They claim that valves offer the advantages of minimal distortion at all levels and the ability to drive all types of loads without risk of damage or instability. Loudspeaker systems, they say, may have an impedance at 50kHz which is many times that of the rating and to provide optimum performance when driving this type of load, a special network is incorporated in the output stage to compensate for this. The first stage of the amplifier comprises a high gain cascode stage which feeds the pentode section of an ECF82 phase splitter. Balanced drive to the output valves is ensured by the setting of a pre-set control in the anode circuits and balanced drive at high frequencies is obtained by an RC network incorporated in the cathode circuit. Stability with any type of load is achieved by limiting the high and low frequency gain by means of an RC network in the anode of the cascode stage. The amplifier, model SA170, has an output of 30W r.m.s. for 0.1% distortion and a sensitivity of 250mV. Frequency response is 20Hz to $20kHz \pm 0.5dB$. Outputs are provided for loudspeakers of 4, 8 or $16-\Omega$ impedance.

A complementary control unit for the SA170 amplifier is the all-transistor SCP1 which has a sensitivity on disc of 2mV with a noise figure better than 70dB. Inputs are provided for disc, tape, microphone and two auxiliary (one 50mV at

Richardson SA170 stereo power amplifier.





Circuit of the Richardson SA170 showing the method of balancing the output valves.

 $2M\Omega$ and the other 250mV at $100k\Omega$). The unit contains its own power pack and has two switched mains outlet sockets for feeding a turntable and the power amplifier. The front panel features a phase-reversing switch which operates on the left-hand channel only to simplify speaker phasing and compensate for recording errors. Price of the SA170 amplifier is £60 and of the SCP1 control unit £45. J. Richardson Electronics Ltd., 43B Hereford Road, London W.2. W.W. 334 for further details.

Slot Stereo

An American tape transport mechanism employing eight-track endless loop recorded cassettes forms the basis of a tape player system new to the British domestic scene, introduced by Slot Stereo. The cassettes employ quarter-inch tape that is pulled past the playback heads from the centre of a reel and re-wound on the outside of the same reel as is standard practice in endless loop cassette systems. To reduce friction the back of the tape is coated with a thin lubricating film of graphite. Track changing is achieved by moving the heads with a solenoid and cam system actuated by a length of aluminium foil spliced into the tape. In this way the track is changed once every revolution of the tape loop so that each of the four stereo tracks is played in turn before being automatically repeated. Tracks can also be switched manually. The player employs a 12-V d.c. motor that is governed electronically by measureing the back e.m.f., which is a function of speed, and comparing this with a reference voltage. Any difference between these voltages results in an error signal which is used to control motor speed.

Three versions of the tape player are available differing in application and cost but employing the same basic tape mechanism.

The car unit measures $8\frac{1}{2} \times 3 \times 7\frac{1}{2}$ inches $(21.5 \times 7.6 \times 19 \text{ cm})$, requires a 12-V positive earth supply at 1.5A, uses 14 transistors, delivers 4W r.m.s. output per channel with a 40dB signalto-noise ratio and costs 47gn.

The 'satellite' unit for home use in conjunction with an existing amplifier system has a performance similar to that of the car unit except that it uses a mains supply and has a low level output only. It costs 47gn.

The home unit incorporates twin 8W r.m.s. output amplifiers but except for the case is the same as the 'satellite' unit, and costs 85gn.

A large variety of recorded music is available in the eight-track cassettes. Incidentally, the R.I.A.A. track pairing standard is employed, i.e. 1 with 5, 2 with 6, 3 with 7, and 4 with 8. We understand



The domestic Slot Stereo casette tape player with (above) the 'satellite' unit.

that Slot Stereo intend to introduce a radio tuner that will slot into the tape players in place of the tape cassettes! Slot Stereo Ltd., 36 Hertford Street, London W1.

W.W. 314 for further details.

Grundig Stereo Tuner/ Amplifier

Five wavebands and 40W of output power (continuous rating) are provided by a new Grundig stereo tuner/amplifier model RTV600 which is fully 'integrated' with pre-amplifiers and control unit. A total of 53 transistors and 31 diodes are used with f.e.ts featured in the front end. On the f.m. range (87-108MHz), six separately tuned programme selectors allow five v.h.f. stations to be pre-set, the sixth selector being for manual tuning. On a.m., the waveband coverage is 145-350kHz, 510-1,620kHz, and 3.15-22.5MHz in two bands. A s.w. fine-tuning control has a range

170



of ± 60 kHz. Sensitivities are 1.5μ V for 15kHz deviation (f.m.) and $6.5-10 \mu$ V through the a.m. ranges. Frequency drift is 1kHz/°C which is compensated by a.f.c. with a capture range of ± 250 kHz. Signal-to-noise ratio is 68dB at 21W, output deviation at linear frequency response measured from aerial to mono loudspeaker output. Pilot tone suppression is 40dB at 19kHz (55dB at 38kHz). A mono/stereo selector is incorporated in the integral decoder controlled by the input level, and stereo channel separation is better than 40dB at 2kHz measured from the aerial input to the loudspeaker output.

In the amplifier section frequency response is 50Hz-16kHz \pm 1dB. Switchable inputs are provided for magnetic or crystal pickups, with preamplifier and correction networks. The output may be connected to loudspeakers with impedances between 4 and 16Ω , the full rated power is available for impedances between 4 and 7 Ω . The output transistors are protected against overload by an automatic circuit which disconnects the amplifier channel if the load exceeds the maximum permissible figure. In addition, an excess temperature switch disconnects the amplifier if the maximum permissible temperature is exceeded. In both cases, the tuner will revert to correct operation when the source of the excess load or the excess temperature has been removed. Controls include a mono/stereo pressbutton selector, stereo balance control, scratch filter and treble and bass controls. The tone controls have a range of -20dB to +15dB at 20kHz and 20Hz respectively. When they are set at zero, the response is linear. Dimensions are approximately $60 \times 15 \times 30$ cm. Price: 198gn. Grundig (GB) Ltd., Newlands Park, Sydenham, London S.E.26.

W.W. 333 for further details.

2 cu.ft Loudspeaker System

The Model DM3 loudspeaker system shown by Bowers & Wilkins is characterized by having an enclosure of moderate size— 0.057 m^3 (2 cu.ft) —and by a new design of bass unit which has an elliptical cone of laminated structure. This bass unit, developed by E.M.I., measures 34 ×

Laminated cone material used in the Bowers & Wilkins loudspeaker.



20.5cm $(13\frac{1}{2} \times 8\frac{1}{8}in)$ and the laminated cone material consists of a mesh of fibre glass sandwiched between two layers of other fibrous material (see photo). Middle frequencies are handled by a pressure unit operating from 3kHz to 14 kHz and the remainder of the range, 14kHz to 25kHz, by a high-frequency unit. Frequency response is better than \pm 2dB from 60Hz to 14kHz and better than \pm 5dB from 40Hz to 25kHz. The cabinet, measuring 72.5 \times 40.3 \times 29.5cm, is a vented type and contains a membrane absorber and wool fibre. Power handling capacity is 15 watts continuous (25 watts music power). Distortion, with 10 watts continuous input, is: 60Hz, 3%; 200Hz, 2.5%; 1kHz, 1.2%; 5kHz, 1.0%; 10kHz, 2%. Bowers & Wilkins Electronics Ltd., Worthing, Sussex.

W.W. 326 for further details.

Ferrograph Series Seven

The gleaming extruded aluminium trim and polished teak cases of the new Ferrograph range of tape recorders is a welcome change after the austere black and grey that has characterized this company's products for the past eighteen years. The models making up the new range, known as series seven, consist of a number of options based on the same design. Facilities available are mono half-track, stereo half track, stereo quarter track, three tape speeds, 19 to 4.75 cm/sec or 38 to 9.5 cm/ sec, with or without output stages and choice of cabinet-portable or natural wood or, if desired, the machine will be supplied uncased for inclusion in an existing cabinet. The basic models are made up from various compatible combinations of these options and range in price from about f_{135} to f_{185} . The new deck has a nominal speed accuracy of 1% and a wow and flutter figure varying from 0.08% at 38 cm/sec to 0.2% at 4.75 cm/sec.

To reduce intermodulation distortion, f.e.ts have been used in all low-level amplifier input stages resulting in a total distortion figure of 0.25% r.m.s. at output levels up to 10W. At the lowest tape speed the frequency response is 50Hz to $7kHz \pm 3dB$, this improves to 30Hz to 20kHz± 2dB at the highest tape speed. The signal-tonoise ratio, unweighted and including hum, is better than 55dB and track separation on stereo models is approximately 45dB. To fully modulate the tape a signal at the microphone input should be greater than 150 μ V but less than 15mV into $10k\Omega$ or more, the recommended source impedance lies between the limits 250 to 2000Ω . The line input is $2M\Omega$ and requires an input of between 75mV and 10V, in this case the source impedance is non-critical. The two 6×4 -inch (15 × 10cm) elliptical speakers incorporated are of course intended purely for monitoring purposes and the loudspeaker output socket provides up to 10W r.m.s. into speakers of 8-16 Ω . A low-level

output per channel is also provided and will deliver 300mV into a load of not less than $10k\Omega$.

Other features include variable speed rewind (max 1min./1200ft), recording level meters switchable to read bias with an easily accessible bias adjustment, tape/original switching through to output stages, three motors, remote control facilities, provision for fitting signal operated switching units etc. Ferrograph Co. Ltd., 84 Blackfriars Road, London S.E.1.

W.W. 328 for further details.

"Free field" magnetic cartridge

By using a fixed magnet and a moving lightweight tube of magnetic material in their '800' stereomono pickup cartridge, Goldring have reduced the mechanical impedance of the moving system and so raised the resonance point to a frequency above 20kHz. Recommended tracking force is 1-3 grammes, but in a version with greater reduction of mechanical impedance, the '800/E', it can be down to $\frac{3}{4}$ gramme. In the '800', which has a diamond stylus with a conical point, the effective ip mass is 1mg and the static compliance 2 cm /N (20 × 10⁻⁶ cm/dyne). Sensitivity is 1mV/cm/s and frequency range 20Hz to 20kHz. Separation does not fall below 15dB and averages over 20dB at



middle frequencies. Performance of the '800/E', which has an elliptical diamond stylus, is somewhat higher. Both cartridges have Mumetal magnetic shields and removable styli. Goldring Manufacturing Co. (Great Britain) Ltd., 486-488 High Road, Leytonstone, London E.11. **W.W. 321 for further details.**

Semi-professional Recorder

The Philips semi-professional recorder PRO12 arrived in this country just in time to be shown at the Audio Festival. Unfortunately, full technical details of some of the more novel features did not. However, sufficient information has been gleaned to make a description worthwhile. The machine has two speeds, 19 and 9.5 cm/sec, which can be adjusted within 0.8% by the eddy current speed control system employed. This is used in conjunction with a built-in stroboscope. In order to achieve a smooth fast wind and to ensure the minimum of wow and flutter and mechanical noise during record and playback the left hand tape guide rotates during fast wind but is locked during other functions. To assist in maintenance the complete head block is removable. Wow and flutter depends on the speed selected but does not exceed 0.13% peak, fast wind time is less than 75 seconds per 1600m of long-play tape and the maximum spool diameter the machine will accommodate is 7 inches (18cm).

Normally the machine is supplied for two-track stereo or mono operation but four-track heads



Philips semi-professional recorder PRO12.

can be supplied as an optional extra if required. The frequency response is to DIN45511 standard and is within 2.5dB from 60Hz to 18kHz (1.5dB from 60Hz to 12kHz) at 19 cm/sec in the playback mode only. The overall record /playback response increases the limits to 5 and 3dB. Signal-to-noise ratio, weighted and measured according to DIN45405 at 19 cm/sec is - 56dB and cross talk rejection is better than 52dB. Crosstalk was measured at 1kHz (3% harmonic distortion) by recording one track at full level and h.f. bias on the second track. The second track is then replayed and measured. The record and playback amplifiers introduce less than 0.5% harmonic distortion. Three inputs per channel are provided as standard and two others may be included if required. They are (1) line, 100mV, 100k Q, (2) microphones, 1mV (unbalanced) suitable for microphones from 50 to 200Ω , (3) diode, 2mV, $20k \Omega$. The optional inputs are (4) microphone with transformer (balanced) for input -50Ω and 0.4 mV—200/500 Ω , and (5) 0.2 mVtransformer line input 0.775V, $10k\Omega$. Each channel has a $10k\Omega$ line output with a nominal output voltage of 775mV, 4V max., in addition a line output transformer can be supplied providing a nominal 0dBm (6dBm peak) output. A stereo monitor output socket is also fitted and has the same characteristics as the standard line output. The single monitor amplifier provides 500mW into the internal speaker at 1% distortion. Twin Vu meters are employed and together with their associated amplifiers are within 2dB. These operate with a 0-10dB switch which is used for setting up purposes. Philips Electrical Ltd., Century House, Shaftesbury Avenue, London W.C.2.

W.W. 317 for further details.

Philips Audio Plan

Three new units which now form part of the "Philips Audio Plan", the method by which a selection of basic units can be formed into a combination according to price or other requirements, are the stereo a.m./f.m. tuner GH944 and two stereo amplifiers GH943 (2 \times 7W output) and GH949 (2 \times 20W output). The GH944 covers the v.h.f., long-, medium-wave bands and short-wave bands from 16.48-50.8m. It has a built-in stereo decoder with automatic switching and stereo indicator. Audio response on f.m. is from 20Hz to 15KHz \pm 1.5dB with standard de-emphasis applied. Sensitivities are 6μ V f.m. mono (60μ V stereo) and 100μ V on a.m.

Stereo amplifiers GH943 and GH949 are both designed to work into an $8-\Omega$ load. The GH943 7W (continuous) model employs 10 transistors and diodes while the higher output model GH949 employs 22 transistors and diodes. Both are equipped with rumble and scratch filters, treble and bass tone controls, and inputs for pickup, tuner and tape. All units are for 110-240V, 50-60Hz, mains operation. Dimensions: $38 \times 22 \times 13$ cm (GH944), $34 \times 21 \times 14$ cm (GH943), and 34×10^{-24}

27 × 14cm (GH949). Prices have not yet been announced. Philips Electrical Ltd., Century House, Shaftesbury Avenue, London W.C.2. W.W. 336 for further details.

Five-unit Loudspeaker

Five drive units are used in the Ditton 25 loudspeaker system, introduced by Celestion, which is stated to be of studio monitoring quality. For the bass there is a 12-inch (30-cm) unit plus an auxiliary bass radiator of the type used in other Celestion speakers. For the higher ranges there are two mid-frequency pressure units and one high-frequency pressure unit. Frequency range is 20Hz to 40kHz and power handling capacity is 25 watts continuous (50 watts peak). Impedance is 4-8 ohms. Priced at £55 13s, the loudspeaker has an enclosure measuring 81×36



Ditton 25 loudspeaker showing the five units.

× 28cm with either teak or walnut finish. Rola Celestion Ltd., Thames Ditton, Surrey. W.W. 327 for further details.

Sinclair System 2000

A complete audio system was introduced and demonstrated at the Festival by Sinclair Radionics, comprising an f.m. tuner, a 35-W integrated stereo amplifier and a plinth-mounted circular loudspeaker. Measuring $30 \times 15 \times 5$ cm the f.m. tuner employs silicon transistors and features inter-station muting, fine tuning, a.f.c. and a pulse-counting discriminator. This type of discriminator has been found by the makers to provide better linearity and freedom from distortion than the ratio and Foster-Seeley discriminators. The tuner is available with or without stereo-decoder which can be plugged in to the mono version to convert to stereo when

The f.m. tuner of the Sinclair 2000 system.



Sinclair cylindrical loudspeaker.

required. Stereo /mono switching is automatic and a stereo neon indicator is incorporated. The use of a variable capacitance diode makes remote tuning possible by means of a plug-in variable resistor, and a switched tuning module designed to fit the same socket will shortly become available. Frequency coverage is 86-108MHz and output 0.5V low impedance. Frequency response is quoted as 10Hz to 20kHz \pm 1dB. Built-in power supplies enable the tuner to operate from 200-250V, 50-60Hz a.c. The entire case and all controls are constructed from solid aluminium. Price 29gn. (25gn mono).

System 2000 amplifier also measures $30 \times 15 \times 5$ cm and is specified as having an output of 35W r.m.s. into a load of 3 to 15Ω . The specification also includes total harmonic distortion <0.5% at 1kHz and full output power, frequency response 15Hz to 30kHz ± 1 dB at 1W and a damping factor of 50 at 1kHz. A transformerless circuit is used employing 12 silicon and 6 germanium transistors with 40dB negative feedback applied to the power amplifiers. Input facilities are fitted for the connection of pickups (magnetic and ceramic), tape (19 and 9.5cm/s), radio, microphone and auxiliary, with input selection by press-button. Price 29gn.

Housed in an aluminium cylinder and priced 12gn. the System 2000 louds peaker introduced by Sinclair Radionics has a frequency response of \pm 2dB from 100Hz to 15kHz and \pm 5dB, 50Hz to 18kHz. Its power handling capacity is 10 watts continuous (20 watts music). The impedance is 8 ohms at 1kHz. The cylindrical enclosure has a diameter of 216mm ($8\frac{1}{2}$ in) and depth of 102mm ($4\frac{1}{2}$ in). Overall height of the speaker on its stand is 380mm (11in). Sinclair Radionics Ltd., 22 Newmarket Road, Cambridge.

W.W. 320 for further details.



Recorder with F.E.T. Input Stages

Just introduced, the CBL/7T retains a distinct similarity, appearance wise, to earlier Vortexion recorders. The black case is retained and the amplifiers for each channel are mounted either side of the tape deck as was the previous practice. The mechanical performance is identical to that of the Ferrograph as both employ the same model seven tape transport by Wright and Weaire, wow and flutter less than 0.1%, 8.25 inch spools (with lid closed), spool height and adjustment, three speeds and provision for remote control.

Eight 1.e.ts are employed in amplifier input stages operating under "starved input" conditions in order to reduce intermodulation distortion to a minimum (0.1% at 10W). These input stages are individually mumetal shielded and are then further shielded by a steel screen to achieve the 58dB signal noise ratio. The f.e.ts employed are SGS-Fairchild types with 0.1dB noise figure.

The complementary symmetry power amplifiers provide 10W r.m.s. per channel output and employ a similar overload protection circuit to that described by Dr. Bailey elsewhere in this issue. The amplifiers are not new, however, as they have been used in other Vortexion equipment over the past two years and are therefore proven from the reliability point of view. The tone control circuit employed is a modified Baxandall arrangement which provides up to 13dB bass and treble lift and cut. This allows for reasonable tonal balancing at the lower output levels.

The low impedance microphone inputs are matched to the high input impedance of the f.e.t. first amplifier stages in mumetal screened transformers, $40-50\mu$ V is required at this input to fully modulate the tape. The line input has an impedance of $100 \text{k} \Omega$ and requires 40 mV running into overload at about the 1V input level.

At a tape speed of 19cm/sec the overall fre-



Vortexion CBL/7T tape recorder.

quency response is within $\pm 2dB$ from 30Hz to 18kHz, however, the makers say that this is conservative and would typically be within $\pm 1dB$.

Other features of the recorder, which costs \pounds 189, include separate speaker volume control so that the monitoring level can be adjusted independently of record level, "before and after" monitoring, bias measuring facility, cross mixing of channels and an echo facility. Vortexion Ltd., 257/263 The Broadway, London S.W.19. W.W. 329 for further details

Kirksaeter Products

Although this German company has been exporting to Common Market countries for a number of years it is a new name in the British hi-fi market and one which appeared at the Audio Festival for the first time this year. Their demonstrations highlighted in particular various versions of integrated tuner/amplifiers. In these the collectors of the Class B output transistors are connected in a bridge circuit which results in a very low internal resistance so that the output is not significantly affected by different speaker impedances. The units are fitted with one loudspeaker



Kirksaeter RTX400 stereo f.m. receiver-amplifier.

output only to which it is permissible to connect a speaker of any impedance between 4Q and 6Q. The RTX400 employs 44 silicon transistors and 16 diodes and provides a sine wave output of 65W. Radio covers the f.m. (87.5-108.5MHz) band only with automatic mono/stereo switching. Inputs are provided for both German (DIN) and U.S.-type sockets and a rumble filter or scratch filter can be switched-in by press-buttons. Model RTX700 has a similar specification to the RTX400 except that the power output is higher at 100W sine wave. U.K. Agents: A.C. Farnell Ltd., 81 Kirkstall Road, Leeds 3, Yorkshire.

W.W. 338 for further details.

Two-unit microphone

A narrow, cylindrical moving-coil microphone which looks more like a capacitor microphone is the D224 from A.K.G. (shown by Politechna). This microphone, which has a cardioid characteristic, contains two capsules, one for the highfrequency part of the range and one for the low-frequency part, and a cross-over network. In this respect it is similar to the earlier D202 which was introduced last year (see W.W. April 1968, p. 59, for explanation of principle). It has a frequency range of 20Hz to 18kHz, a sensitivity of -77dB relative to $10V/N/m^2$ (0.15V/ μ b) and an impedance of 200 ohms at 1kHz. Politechna (London) Ltd., Eardly House, 182/4 Campden Hill Road, Kensington, London W.8. W.W. 322 for further details.

Record Reproducer Plinth

For housing their pickup arms, and various makes of turntables, S.M.E. have introduced a record reproducer plinth constructed in rose-wood with a one-piece acrylic lid. It features a four-point spring suspension system with adjustable damping which carries the motor board and protects it from acoustic feedback and external vibration. A range of interchangeable motor boards wil be available, ready cut and drilled to accept various combinations of arms and turn-tables. Price is $\pounds 25$ 4s. S.M.E. Ltd., Steyning, Sussex.

W.W. 325 for further details.

S.M.E. record reproducer plinth with acrylic lid.



Belt-driven turntable

A flexible belt-driven system is used to reduce vibration and transmission noise in the Connoisseur BD1 turntable, which is intended for high-quality sound reproduction but has a relatively low price (£13 19s 2d). The turntable proper is a 26cm ($10\frac{1}{2}$ in) diameter lathe-turned aluminium casting with a phosphor bronze bearing. Drive is from a slow speed synchronous



Connoisseur turntable and pickup.

motor. The speed required $(33\frac{1}{3} \text{ or } 45 \text{ r.p.m.})$ is selected by lifting a hinged metal cover and moving the ground rubber cord drive belt. Rumble is -60dB and hum -80dB relative to 7cm/s at 1kHz. Wow and flutter are approximately 0.1 of 1%. The turntable is available in chassis form or on a plinth with a pickup mounted. A. R. Sugden & Co. (Engineers) Ltd., Market Street, Brighouse, Yorks.

W.W. 340 for further details.

Kit Tape Recorder

Designed to satisfy the needs of those who prefer to build their own equipment this particular member of the Daystrom kit-family has three speeds, four tracks and incorporates a total of eighteen transistors. The BSR TD10 tape deck supplied with the kit will accommodate up to seven-inch spools, incorporates a digital tape position counter and at the highest speed (19cm/sec) the wow and flutter figure is 0.15%. Power amplifiers provide 4W r.m.s. per channel into the two internal 8 \times 5 inch (20 \times 13cm) loudspeakers; unweighted signal-to-noise ratio is better than 40dB. Again at the highest speed, the frequency response is within $\pm 3dB$ from 40Hz to 18kHz and at the lowest speed (4.5 cm/sec) the upper 3dB limit falls to 7.5kHz. Two inputs are provided per channel and are: microphone, requiring $350\mu V$ into $50k\Omega$; and auxiliary, needing 50mV into 80k Ω . Likewise two outputs per channel are provided, 4W into 15Ω and 250 mVfrom $4k\Omega$. The recorder, which costs £58 in kit form or £70 assembled, comes complete with a moving coil microphone, 5.75 inch (15cm) reel of tape and the sort of comprehensive instruction manual for which Heathkit have become wellknown. Daystrom Ltd., Bristol Road, Gloucester. W.W. 315 for further details.

Truvox Series 200

The six models that comprise the 200 range of tape recorders consist of two- and four-track stereo and mone machines available in a choice of p.v.c. or teak cases.

The tape deck used is the same on all machines and has seven-inch (18cm) reel capacity, three speeds, 19, 9.5 and 4.75 cm/s with a wow and flutter content of 0.25% for the lowest speed and 0.1% for the highest. It incorporates three motors, has a rewind time for 1200ft of 2 min and has an outside rotor capstan motor. All four-track models have an unweighted signal-to-noise ratio of 48dB and the two-track models 50dB. The frequency response is the same in each model and is 30Hz to 18kHz \pm 3dB at 19cm/sec and 40Hz to 7kHz \pm 3dB at 4.75cm/sec. Two inputs per channel are provided, the low-level microphone input requires not less than 1mV into 50k Ω and the line input needs 50mV into 200k Ω .

Mono versions cost in the region of £125 and include a 10-W output stage and an 8×5 inch $(20 \times 13 \text{ cm})$ hyperbolic cone loudspeaker. A semi-cardioid moving-coil microphone is also sup-



One of the Truvox "200" tape recorders.

plied. Stereo versions are intended for building into a system so do not incorporate output stages or loudspeakers. These cost about $\pounds 150$. Truvox Ltd., Shore Road, Hythe, Southampton. W.W. 313 for further details.

Braun Stereo Amplifiers

Two new Braun stereo amplifiers were demonstrated by Fi-Cord International, both types being fully transistored. The CVS250 is a 2 x 15W (continuous) amplifier which is matched in design and colour scheme to the Braun CE500 tuner. It is equipped with a volume control with push-pull switch for by-passing loudness compensation, separate bass and treble controls for each channel and a stereo /mono switch. Frequency response is said to be 30Hz to 30kHz and distortion < 0.5% at 12W. Inputs are provided for radio, gram, tape and reserve, and outputs for two 4- Ω impedance loudspeakers. Dimensions: $26 \times 11 \times 32$ cm. Model CSV500 is a combined pre-amplifier and power amplifier with switchable rumble filters, treble filters and presence boosting. It has a power output of $2 \times 45W$ (continuous) for $4-\Omega$ loudspeakers and slightly less power when connected to 8Ω loudspeakers. Distortion is given as $\Omega 0.5\%$ (4 Ω) or 0.3% (8 Ω). Separate bass and treble controls are provided for each channel. Frequency response is quoted as 10Hz to 35kHz. Five inputs are provided for gram,



Braun CSV500 stereo amplifier and pre-amplifier.

tape, radio, microphone and reserve, and outputs for loudspeakers of 4-16 Ω and headphones of 400 Ω impedance. Dimensions: 40 × 11 × 32cm. Prices: CSV250, £103 .5s.; CSV500, £203 10s. U.K. agents Fi-Cord International, Charlwoods Road, East Grinstead, Sussex. W.W. 331 for further details.

Loudspeaker kit

At a price of $f_{.10}$ 10s Rank Wharfedale have introduced a high-fidelity loudspeaker kit containing an 8-inch (20-cm) bass /middle-range unit, a pressure unit for the treble, a cross-over network, connecting wire, mounting bolts and acoustic wadding. The 8-inch unit has a magnet with a field strength of 9.6 \times 10 5 A/m (12,000 oersteds), flux of 4.8 \times 10 $^{-4}$ weber (48,000 maxwells), a die-cast chassis and a flexible roll surround. Its free air resonance is 35Hz. The pressure unit has a magnet field strength of 8.4 imes 10⁵ A/m (10,500 oersteds) and has a dome diaphragm with a diffuser cap. Frequency range is 40Hz-17kHz and the cross-over frequency is 1,750Hz. Power handling capacity of the whole system is 15 watts and impedance 4-8 ohms. Cabinets (parts not supplied) can be any size between a minimum 356 × 248 × 223mm (bass restricted to 65Hz) and a maximum 293 \times 560 × 242mm. Rank Wharfedale Ltd., Idle, Bradford, Yorks.

W.W. 319 for further details.

Record Cleaning Machine

Intended for use by public libraries, record dealers, broadcasting organizations etc., a disc record cleaning machine introduced by Audio & Design comprises a turntable, a liquid application system and a suction cleaning mechanism. The turntable carrying the record revolves at high speed and a liquid solvent is applied by swinging a brush over and down onto the record. Considerable pressure can be applied since the disc is well lubricated by the liquid. This action dissolves grease and loosens grit on the record. Next a suction arm is moved to the centre of the disc. It is held by suction to the surface, and a tracking mechanism moves it outwards towards the periphery of the revolving record. This removes the liquid with its dissolved and suspended foreign matter and leaves the record dry and free of chemicals. The whole process takes about two minutes per side. Audio & Design (Recordings) Ltd., Special Products Division, 40 Queen Street, Maidenhead, Berks.

W.W. 324 for further details.

Fisher Hi-Fi F.M. Receivers

A new ittem added to the well-known American Fisher range seen at the Audio Fair was a high-quality f.m. receiver with pre-set press button tuning. This tuning arrangement is one which has appeared for some time on a number of German f.m. tuners and a Fisher representative explained that a German-made tuning device was, in fact, incorporated in the design. Fisher call their device "Tune-O-Matic" and it comprises five tuning knobs, each knob having its associated vertical frequency scale and a cursor on the receiver facia. Each knob can be tuned to a different f.m. station and pre-set, so that subsequent changing from one station to another only requires the appropriate tuning button to be depressed. Stations can also be tuned manually if desired. The receiver, model 160-T, provides r.m.s. power of 15W per channel at 1kHz with 0.5% harmonic distortion. Audio bandwidth is 25Hz to 25kHz. Usable sensitivity of the tuner section is $2.2\mu V$ and signal-to-noise ratio at 100% modulation and 1mV input is 60dB. The design incorporates i.cs, f.e.ts and a silicon-transistor complementary output stage



Fisher 160-T 40-watt f.m. stereo receiver.

with short circuit protection. The presence of a stereo broadcast automatically switches the decoder to stereo and operates the indicator lamp. Tone controls are of the Baxandall design. Dimensions are $39 \times 8 \times 29$ cm; weight 6.8kg and power consumption 65W. Price: 129gn. U.K. Distributors: Getz Bros. & Co. Inc., 2 Harewood Place, London W.1.

W.W. 332 for further details.

Noise cancelling microphone

The type 4-70 noise cancelling ribbon microphone just introduced by Lustraphone uses two ribbons, inclined towards each other like the sides of a 'V' and connected in antiphase. When the direction of the sound waves is such that they enter symmetrically between the two ribbons (into the top of the 'V') the ribbons move in opposite directions, and because they are connected in antiphase their signals are additive. Sound waves arriving at either side of the 'V', however, cause both ribbons to move in the same direction, and in this case the two electrical signals are subtractive.



This results in an acceptance angle of about 50° for wanted sounds, while unwanted sounds those arriving from all other directions outside this acceptance angle—produce very weak signals. Price of the 4-70 microphone is £26 5s. Lustraphone Ltd., St. George's Works, Regents Park Road, London N.W.1. W.W. 323 for further details.

Sansui Control Amplifier

One of the exhibits from Japan was a solid-state combined stereo control and power amplifier, model AU-777 by Sansui. This unit has a continuous power output rating of $2 \times 30W$ and a sensitivity of 2mV for rated output. Harmonic distortion is <0.5% and power bandwidth (i.h.f.)



Sansui AU-777 control amplifier with a continuous output of 25W per channel.

is 20Hz to 50kHz. Frequency response of the main amplifier is given as 20Hz to 100kHz ±1dB at normal listening level. Dual concentric negative feedback type tone controls give treble and bass lift and attenuation of 15dB. The pre-amplifier and main amplifier sections are designed so that they can be used separately in conjunction with other equipment. A total complement of 26 transistors and 13 diodes is used in a transformerless circuit with protection against accidental shortcircuits. Four outputs and seven inputs provide the AU-777 with all the necessary inputs and outputs for an amplifier: pre-amp, tape, and two centre channel outputs, two gram inputs, tape and auxiliary inputs. The centre channel outputs are two terminals for connection to a third mono amplifier which is said to give a three-dimensional effect. The full control range includes a loudness control and presence switch, also a speaker selection switch. Price 105gn. U.K. Agents: Technical Ceramics Ltd., Cheney Manor Industrial Estate. Swindon, Wiltshire. W.W. 337 for further details.

Rogers F.M. Tuner

Provisional details were available at the Audio Fair of a new f.m. tuner unit being displayed by Rogers Developments which they call the Ravensbourne 2. Although designed for use with the Ravensbourne stereo amplifier, it is suitable for use with many other amplifiers. The Ravensbourne 2 is an example of a product designed primarily for export, but which is recommended for use in this country where reception conditions are particularly difficult. It has high sensitivity $(3\mu V \text{ for full limiting})$ and high selectivity, but adequate bandwidth for stereo reception. Three f.e.ts are featured in a four-gang front end followed by six silicon and one germanium type. An optional decoder contains a further seven silicon transistors. Audio output is 150mV r.m.s. average (mono) and 200mV r.m.s. average (stereo). Four different prices are quoted for the tuner according to whether it is supplied with or without decoder and case. The top price for a complete stereo model with case is $\pounds 63$ 10s and its dimensions are $32 \times 29 \times 13$ cm. Rogers Developments (Electronics) Ltd., 4-14 Barneston Road, Catford, London S.E.6.

W.W. 339 for further details.

Ravensbourne 2 f.m. tuner (Rogers).



Wireless World, June 1968

Arena Tuner/Amplifier

Made by Hede-Nielsen Fabriker, of Denmark, the Arena products stand included tuner/amplifier model T1500F which became available in May. This is a stereo a.m./f.m. tuner combined with power amplifiers which provide 6W output per channel. 87-104MHz is covered on the f.m. range, and on a.m. a s.w. band covers 40-51m plus the usual medium- and long-wave broadcast bands. Sensitivity on f.m. is $1\mu V$ and on a.m. $10\mu V$. Socket facilities provide for up to four $4-\Omega$ impedance loudspeakers, magnetic (6.5mV at $50k\Omega$) and crystal (450mV at $5M\Omega$) pickups, and tape recorder input and output. Separate bass and treble combined tone controls give ± 15 dB and ±10dB variation respectively. Frequency response is claimed to be 20Hz-18kHz and harmonic distortion 0.15% at 5W. The T1500F measures $15 \times 42 \times 21$ cm and employs 20 transistors and 20 diodes. Special features include an automatic f.m. scanning system and preselection of five f.m. stations. Price 61gn. Arena products are distributed in the U.K. by Highgate Acoustics, 184-188 Great Portland Street, London W.1. W.W. 330 for further details

Books Received

F.E.T. Applications Handbook edited by Jerome Eimbinder. Introductory chapters look at the various types of f.e.t. that are available, discussing their electrical characteristics and applications. Two chapters are devoted to biasing problems and techniques and a further chapter concentrates on the device as a constant current source. Oscillators, noise, audio amplifiers and pre-amplifiers, v.h.f. pre-amplifiers, source followers, phase splitters and drive circuits occupy a further seven chapters. The f.e.t. as a switch forms the subject heading under which the majority of the remainder of the book falls and the f.e.t's use in chopper, integrated and commutating circuits is described as are serial logic arrays. The final chapter is concerned with the photo-f.e.t. Pp 286. Price \$12.95. Tab Books, Blue Ridge Summit, Pal7214, U.S.A.

F.E.T. Circuits by Rufus P. Turner contains a wide range of practical circuits. Some difficulty may be experienced by the English reader because of the American terminology used, however, as with other books from this publisher, a section is included to assist the English, and this together with a little common sense should surmount any problems which may occur. The first chapter is devoted to the construction, operation and performance of the f.e.t. Constructional projects are under six headings: amplifiers, oscillators, receivers, transmitters, control and instrument circuits. The amplifier section includes circuits for a variety of audio amplifiers, and an audio notch filter, and proceeds up the frequency spectrum through i.f. to video amplifiers. The oscillators described similarly cover the frequency range from audio to r.f. The transmitter section includes a number of designs but the English reader should tread with caution in view of the difference in licence regulations between here and America, a subject which is discussed in the advice to English readers' section. Control circuits cover the usual range of devices in the electronic novelty category: timers, sound operated relay, etc. The instrument section describes about twenty assorted electronic test instruments. Pp 160. Price 26s. W. Foulsham & Co. Ltd., Slough, Bucks.

Transistor Audio Amplifiers by S. J. Hellings, from the Philips Technical Library, is roughly split into two sections; the first covering transistor theory and the second amplifier design. The transistor theory starts with the necessary basic description of the structure of matter and proceeds to the relationships between the *h*-parameters and transistor properties. The remainder of the book (21 chapters) is devoted to amplifier design problems —feedback, temperature, frequency limitations, interstage coupling, tone control, equalization, noise, etc. Several complete amplifier circuits are given together with complete design information. Pp 333. Price 104s. Macmillan & Co. Ltd., Little Essex Street, London W.C.2.

Practical Television Circuits by R. E. F. Street. There are enough circuits in this book to keep the most industrious constructor going for years. Constructional information is given on five aerial pre-amplifiers, the Olympic II transistor 14-inch television receiver (405 lines only), a TV hearing aid and a baby alarm, interference suppression circuits, TV and converter and test equipment oscilloscope, wobbulator, Q-meter, e.h.t. voltmeter, signal strength meter, etc. Pp 375. Price 30s. Newnes Books, Hamlyn House, 42 The Centre, Feltham, Middlesex.

Fundamentals of Digital Magnetic Tape Units by the Field Engineering Department, Univac Data Processing Division, of the Sperry Rand Corporation. The book should be of value to those schools and individuals who are at the present time employed on a computer project and wish to extend this to embrace magnetic tape storage. Two early chapters look at the various aspects of electricity and magnetism as they apply to tape recording in general and digital tape recording in particular. Twenty pages are devoted to tape transport mechanisms which discuss the various pinch roller, plus vacuum, tape handling methods in use. The recording techniques chapter describes the various ways that are employed to record binary information together with data recovery and error detecting techniques. The last chapter deals with circuits, both read and write, but they are not complete in that head impedances and semiconductor types are not given. Pp 96. Price 21s. W. Foulsham & Co. Ltd., Slough, Bucks.

Semiconductors Vol III-Non-linear circuits by E. J. Cassignol (from the Philips Technical Library). The first part of this volume deals with oscillators, the initial chapter outlining the study methods involved. The oscillator section is split into two, one part covering the harmonic oscillators (phaseshift, tuned circuit, resonator-crystal-tuning fork, etc., bridge oscillators) and the second part relaxation oscillators (various types of multivibrators). Sandwiched between these two is a chapter on various aspects of RC circuits. Thirty pages are devoted to scanning circuits with detailed discussions on the various integration methods that may be employed. The digital circuits chapter includes a summary of Boolean algebra and elementary logic functions and discusses transistor and diode logic circuits, magnetic circuits and binary elements. The remainder of this book concentrates on power supplies. Pp 265. Price 108s. Macmillan & Co. Ltd., Little Essex Street, London, W.C.2.

Television Camera Tubes

Recent developments in the image orthicon

by I. R. Sinclair, B.Sc., M.I.E.E.

In the years after the war, the image orthicon became firmly established as the only television camera tube suitable for a wide range of high-quality transmissions. The hope that the small and comparatively inexpensive vidicon, using a photoconductive layer of antimony trisulphide, would replace the image orthicon, has not been realized. The improved form of vidicon known as the Plumbicon* (devised by the Philips Company, and using a lead oxide layer formed by evaporation of lead oxide in argon gas) has proved a much more likely successor. Its smaller size, its comparative simplicity from the user's point of view, and its good performance make it preferable to the i.o. in many respects, and in colour cameras the problems of obtaining registration from three Plumbicons are much less than the corresponding problems with image orthicons. Nevertheless, to utilize the full resolution capabilities of a colour signal on 625 lines (remembering that the monochrome signal has a much greater bandwidth than the chrominance signals, and is responsible for the impression of resolution seen by the eye) it is preferable to use a $4\frac{1}{2}$ inch i.o. to provide the monochrome signal, and Plumbicons (two or three) for the chrominance signals.

Eventually, the lead oxide tube will probably replace the i.o. completely, but, at the moment its development seems to have slowed down somewhat, while recent work indicates that the full performance of the i.o. has not yet been realized. Economic considerations may also delay the displacement of the i.o. by the Plumbicon; neither form of

* Registered trade mark of N. V. Philips' Gloeilampenfabrieken of Eindhoven, Holland.



After graduating from St. Andrews University, I. R. Sinclair joined English Electric Valve Co. in 1956 as a junior engineer working on direct-view storage tubes. He later transferred to the television camera tube department where he stayed until 1966 when he joined the staff at the Hornchurch College of Further Education. He is now a lecturer in physics and physical chemistry at the Braintree (Essex) College of Further Education

tube is inexpensive, but the replacement of large numbers of cameras and lenses is not a process which is likely to be hurried. When the $4\frac{1}{2}$ -inch i.o. was introduced, it was intended that this technical advance on the 3-inch tube would eventually make the latter obsolete. This has not happened; the lower price of the 3-inch i.o. and its associated equipment has, in fact, until recently caused a steady expansion in the sales of this tube size.

In North America, of course, many commercial stations are run on a budget which allows very little for maintenance or replacement of equipment; similar economic considerations also apply to several European countries. Another factor which has prolonged the life of the 3-inch i.o. in America is the alleged low quality of the picture on domestic receivers; one engineer once remarked that they could get away with a thirty-line system if the wording on the commercials were made slightly bigger!

Operation of the Tube

The tube can be divided, for the sake of understanding its operation, into three sections (see Fig. 1); the image section, dealing with the conversion of the light image into an electron image and the storage of that image in the form of a charge pattern on a dielectric; the scanning section, forming the electron beam which scans the charge pattern and forms a modulated return beam; and the multiplier section, which noiselessly amplifies the modulated return beam to the usable proportions of an output signal at the final anode.

The image section comprises the photocathode (a thin layer of photo-emissive material prepared in situ during the evacuation of the tube), the image accelerator electrode, and the target cup, to which is attached the target and target mesh assembly. When the photocathode, to which an electrical contact is made, is illuminated and made a few hundred volts negative to the other electrodes, electrons leave it in quantities proportional to the strength of the light. The light image formed by the lens on the photocathode has now become an 'electron flow' image, and must be focused by the image accelerator electrode and by an externally applied solenoidal magnetic field to prevent deterioration and to ensure that the size of the electron image matches the size of the target. In the older 3-inch tube, a reduction in size was necessary; in the later $4\frac{1}{2}$ -inch version the image is magnified; in each case this is accomplished by grading the strength of the applied magnetic fields. Modern cameras use additional faceplate coils to improve the geometry of the image section, high (negative) photocathode voltages are also used to minimize chromatic abberration caused by the different initial speeds of electron emission.

The electron flow image is converted to a stored charge image at the target-a thin sheet of dielectric material with a small but important amount of conductivity from front to back. Electrons landing on this target, conventionally made of 0.0001-inch thick soda-glass, charge the surface facing the photocathode positively by secondary emission (each landing electron knocks off more than one existing electron, so the net result is loss of electrons which amounts to a positive charge) as long as the secondary electrons released by this process can be collected at a surface more positive than that of the target face. This more positive surface is provided by the metal target mesh, located about 0.001-inch in front of the target. Besides acting as a collector of secondary electrons, the target mesh serves two other functions. There exists a capacitance between the mesh and the surface of the target (the assembly is in effect a parallel plate capacitor) and it is this capacitance which is used to store charge. The presence of the target mesh also helps to ensure that electrons arriving from the photocathode arrive normal to the target, and not at random oblique angles.

Due to the conductivity of the target, an image of potential appears also on the side facing away from the photocathode, and it is this image which is scanned by the electron beam in the scanning section. The beam, derived from a cathode and grid similar to those used in a c.r.t., is limited in diameter by passing through an aperture of diameter 0.001-0.0025-inch, and is focused electrostatically and magnetically. Electromagnetic scanning is used, and bucking coils are employed in modern cameras to ensure that the scanning field does not extend into the region of the target. In addition, two sets of coils (the alignment coils) whose axes are perpendicular to each other and to the tube's axis, are used to ensure that the beam
approaches the target along lines perpendicular to the target surface. The operating characteristics of the image orthicon are greatly improved if another mesh, the field mesh, is placed between the target and the electron gun at a distance of about 0.25-inch from the target. This has the effect of altering the decelerating field at the target (the mesh is run some ten volts positive to the G4 of the electron gun) and the visible effect is a considerable improvement in black-white transitions, which are very much sharper when a tube incorporating a field-mesh is used.

The target potential at any point is set by the stored charge which in turn is derived from the photocurrent at the corresponding point on the photocathode. The maximum positive excursion of the target potential is set by the voltage applied to the target mesh because secondary electrons can leave the target only if a more positive potential exists in a region which is not co-planar; when a region of the target has the same potential as the target mesh there is no tendency for electrons to travel from one to the other. The minimum potential of the target is set by the beam striking the target-this is equivalent to connecting the target to cathode, and so the minimum target potential is cathode potential. For the best rendering of grey tones by the image orthicon, the target mesh is usually set at 3 volts positive to cathode potential.

As the beam scans the target, varying fractions of the beam current land on the target according to the potential pattern on it. The remaining fraction of the beam is reflected specularly to form a return beam moving back down the scanning section; this return beam is a larger fraction of the incident beam while a portion of target at cathode potential is being scanned, and a smaller fraction of the incident beam during the scan of a region of maximum positive potential. The return beam current is therefore the inverse of the original photocurrent. The forward beam current and the time constant of the target (the product of the through resistance of a unit area and its capacitance to the target mesh) together determine the fraction of the charge at each scan and hence the storage time-the time for which an image is visible after photocurrent to the target has ceased. Since the conductivity of the target varies with temperature, the storage time of an image orthicon is very long when the tube is cold (a phonomenon known as sticking) and short when the tube is hot (causing some loss of sensitivity). In modern cameras, the temperature of the target section of the tube is thermostatically controlled.

The function of the multiplier section is to amplify the return beam signal by a series of secondary emitting stages, or dynodes. As in the case of every amplifier, the first stage is the most important. If the forward beam always travelled exactly perpendicular to the target, the return beam should be the exact inverse of the forward beam. As the angle of incidence varies slightly during scanning, the angle of return also varies, and the effect is that the return beam scans the first dynode. Because of this, any variation of the secondary emission ratio of the first dynode over its scanned area must cause variations in the output signal which superimpose a 'picture' of the first dynode surface on the received picture.

Modern Developments

The Target.-Since the target is the processing element which converts the weak continuous signal from the photocathode into a stronger intermittent (scanned) signal, it is not surprising that a considerable amount of development work has been carried out on this section. Conventionally, the target is made of soda-glass blown into a bubble. A suitable portion of the bubble is selected and sealed onto a metal ring, the rest of the bubble being broken off. A later development is to use two semicircular loops of wire which are electrically heated, coated with glass, and then separated by rotating one loop about the common axis. In this way a circular film of glass is produced, and can be sealed to a metal ring in the same way as before. This method is particularly useful in making the types of glass target referred to later which must be prepared in a non-oxidizing atmosphere.

General Electric of the U.S.A. has pioneered the use of targets of magnesium oxide. This material has a high secondary emission ratio, which has the effect of increasing the sensitivity of the tube, but the resistivity is also very high, which makes it necessary to use very thin films of the material. Although the technology of prepar-

Fig. 1. Sectional diagram showing the disposition of the image, scanning and multiplier sections of an image orthicon



ing thin films of refractory materials is adequate, means of supporting such films are not, and target vibration is even more of a problem than it is with glass targets. Despite this, the life of tubes using magnesium oxide targets is claimed to be very long, the response is linear, and the warm-up time negligible.

These benefits derive from the different conduction mechanisms of glass and magnesium oxide, the former conducting by the movement of sodium ions (since glass is a 'fossilized' liquid, not a solid) and the latter by movement of electrons. Ions are large particles, and their movement in one direction through a glass target depletes one side and enriches the other; there is no mechanism for reversing this process short of melting the glass to permit the ions to move back freely. The consequences of this process are that the resistivity of the target rises steadily during the life of the tube, increasing the time constant of the tube until it is unsuitable for anything but still pictures. The transfer of ions can also take place to such an extent when the camera is left on a stationary scene that the charge image of the scene is permanently 'printed' on the target. These effects limit the life of a tube using the conventional glass target, and the use of electronically-conducting magnesium oxide has shown that image orthicons with very long working lives can be made once this barrier is removed.

Another approach to the desired electronic conduction exists by making glass containing a high proportion of titanium oxide, along with other oxides of the same chemical group. Because these oxides can exist in either of two stable chemical conditions, one being electron deficient, electronic conduction can take place in a glass made of such oxides. This approach preserves the traditional type of material, but imposes severe manufacturing problems, since the glass cannot be worked in the presence of even a minute quantity of air. Targets made from this electronically conducting glass exhibit the same desirable long-life properties as those made from magnesium oxide; and are mechanically more robust.

A third form of target material which is under intensive investigation is the secondary electron conduction (s.e.c.) target. Secondary emitting targets made from aluminium oxide coated with aluminium and with potassium chloride have been known for some time, and are used in light amplifiers as transmission multipliers, that is, multipliers where the secondary electrons emerge travelling in the same direction as the primary electrons causing them. Research by the Westinghouse Co. of the U.S.A. has shown that a small modification to the process of depositing the potassium chloride layer (evaporation in an atmosphere of dry argon) results in a spongy layer with an extremely high secondary emission ratio. Secondary emission ratios of 50-400 seem likely if production difficulties can be overcome; the main difficulty in the use of such targets is that moisture must be totally excluded during their handling. To comply with such a requirement means that the design of an image orthicon must be considerably changed, no conventional glass seal-



Electrode assembly of a $4\frac{1}{2}$ -inch E.M.I. image orthicon television camera tube showing the multiplier and dynodes.

ing operation must, for example, be carried out after target insertion, limiting constructional methods to solder-glass joints or argon-arc welding. With the tube filled with dry argon, the former is unsatisfactory because it requires heating the whole tube; the latter requires considerable redesign of the tube in the target region, though such a redesign would offer an excellent opportunity for rationalizing the design of the target mountings and incorporating improvements in the image section geometry.

Apart from the use of new target materials, the growing problem of target microphony has promoted a large amount of research on the mechanical nature of the processes at the target. Two types of target microphony have been identified; shock microphony, caused by vibration, whether mechanical (in tracking or handling) or acoustical (especially with percussive sounds); and electrical microphony, caused by magnetostriction in the material in which the target is mounted, which starts and stops in random fashion irrespective of the presence or absence of vibration. Any form of microphony is most undesirable in a camera tube, since it is readily visible on domestic receivers, whereas most other forms of signal degeneration (with the exception of sticking) are not.

Both forms of microphony are greatly aggravated by the use of high (3V or more) target mesh voltages, and the increased trouble caused by microphony has followed directly on the adoption of 3V as a standard target potential by major camera tube users because of the better tonal gradation afforded. Shock microphony is in every way similar to the microphony encountered in high-gain amplifier valves, and is dealt with in a similar manner. Movement of the target is equivalent to moving one plate of a charged capacitor, causing a variation in potential at each point of the target. When this occurs during scanning, the potential variations are visible on the monitor picture as horizontal

bands of light and shade moving up or down the picture. The effect is very similar to sound on vision and can easily be confused with it, since the movement and number of the bands varies with the pitch of the sound. One distinguishing feature is that sound-onvision occurs over all the audio frequencies, but shock microphony tends to be confined to the frequencies around the resonant frequency of the target, usually between 350 and 500 Hz.

A cure may be effected by using soft spacing strips between target and mesh, but this causes practical difficulties, since the thickness of the strips must be of the order of 0.001 inch or less, and no projections are permissible. A compromise may be reached where normally constructed tubes are used for the older types of cameras, and specially non-microphonic tubes for cameras which use motor driven lens turrets or motor-operated zoom lenses.

Electrical or "spontaneous" microphony is due to the combined effect of the electrostatic forces due to the charge on the target and the magnetostrictive forces caused by the action of the scanning fields on the metal ring holding the target. The magnetostrictive effect slackens the target so that the magnetic field increases, and the electrostatic field increases the attractive force between target and target-mesh as the target is discharged to cathode potential. The combination of the two causes the target to vibrate at its resonant frequency; the vibration is erratic, however, since the forces on the target vary with the over-all charge and the degree to which the target is discharged.

Factors favourable to spontaneous microphony include high target voltage, low light levels and high scanning field strengths at the target. It is unfortunate that the metal most suitable for sealing to soda-glass, a 50% nickel iron alloy, is also the NI Fe alloy with maximum magnetostrictive effect. The substitution of titanium for nickel-iron in the E.M.I. $4\frac{1}{2}$ inch camera tubes in 1964 effected a complete cure of the spontaneous microphony problem, and no difficulty is presented by the electronically-conducting glasses, since they generally use rings of molybdenum, which also exhibit negligible magnetostriction.

The First Dynode .- The problem of the scanned surface of the first dynode has been tackled electrically and chemically. If the whole gun is tilted (the beam being straightened by a correcting coil) only a part of the surface is in focus for the return beam. A more elegant method is to use a tilted or curved dynode surface, though this causes fabrication and assembly problems. The surface finish and material used for the dynode are both of considerable importance; a smooth matt finish is desirable, and materials include silver-magnesium alloy, chromium and nickel. At present materials may be chosen for long life or for minimum visibility but not both although much was hoped for the nickel dynode at one time.

With the new electronically conducting targets offering very much longer target life-times, the need for a long dynode life to match this has intensified the work being done on the production of new dynode surfaces.

Beam Noise.—The limiting factor in the signal-to-noise ratio of the camera tube is the beam noise. This limitation is aggravated by the tube operating conditions, which cause maximum beam current to return in black areas where beam noise is very visible. The change to 625 lines has made this more obvious, since the signal-to-noise ratio is some 3 to 5 dB lower at the higher scan rate compared with that of 405 lines.

The obvious thing is to reduce the noise of the beam leaving the cathode and this may be done to some extent by using a cathode surface of better finish or of better geometry. A similar problem is encountered with travelling wave tubes, and smooth concave cathodes have materially assisted in reducing beam noise.

A more radical approach is to redesign the tube for Isocon operation. The Isocon principle known for some considerable time but never used commercially until very recently, reduces beam noise by separating off the unmodulated portion of the return beam current and allowing only the modulated portion to reach the first dynode. The modulated portion of the return beam is that portion which is elastically reflected by the target voltage, the remainder of the return beam being electrons which have been scattered at the target, and which, because of their random velocities, constitute a noise signal. In principle, these two components of the return beam can be separated on the basis of their differing lateral velocities by means of deflecting fields and a partition. Working Isocons have been produced, and have a marked superiority in signal-to-noise ratio over conventional image orthicons, but it is doubtful if their very high cost could be justified in broadcast television use, even on 625 lines where the higher scan rate has the effect of reducing the signal-to-noise ratio obtainable from the tube.

There is much work to be done on the image orthicon. The $4\frac{1}{2}$ -inch tube does not have the resolution superiority over the 3-inch which theory would indicate. Some of this lost resolution is due to the image section (more elaborate than that of the 3-inch tube) which suffers from the inevitable lens distortions. Some improvement in resolution (3 dB increase in signal at 5 MHz) can be obtained by quite small changes in electrode shapes in this section. A further loss of resolution arises from the difficulties of maintaining focus and orthogonal target landing over the larger area of the $4\frac{1}{2}$ -inch target; this problem should be tackled by designing the tube and its scanning yokes together. As far as the target region is concerned, the processes of electron distribution which take place at the target are certainly not well understood, yet these processes are fundamental to the operation of the tube. Some of the problems may never be tackled now that all the manufacturers of image orthicons are also making lead-oxide vidicons.

The writer wishes to acknowledge gratefully the help of E.M.I. Ltd. in the provision of information and photographs.

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Demonstrating A.C. Theory

Sampling method using a strobing meter

by J. G. Assenheim, M.Sc., B.Sc., M.I.E.E.

That current should exist in a circuit when the voltage is at zero and vice-versa may be accepted by many students as a theoretical proposition, but it is regarded with scepticism as far as practical circuits are concerned. On first acquaintance, the beginner finds difficulty in appreciating the phase relationships which exist in an a.c. circuit with reactive components. Many lecturers feel that a practical demonstration, to show that these phase relationships do indeed exist, helps to clear up any doubt in the student's mind.

Unfortunately, demonstrations of this type are not easily performed, except on a double beam oscilloscope and, as many second-year students are not familiar with the function of these instruments, they remain unconvinced. Most students approaching a.c. theory for the first time have had experience with, and have learned to trust, the simple moving coil meter. A demonstration showing voltages and currents displayed on centrezero meters placed at various points in a circuit which is energized with a very lowfrequency sinusoidal voltage appears to be the most instructive form of demonstration. A number of colleges provide this type of exercise and typical examples have appeared Wireless World in the past few in years.1,2,3

If the frequency of the source is to be sufficiently low to enable the waveforms to be examined on a number of meters simultaneously, then the component values required are inconveniently large. If the display is not to confuse students, and the lecturer is to be given sufficient time to describe what is happening, the frequency should not be higher than about 0.2Hz, i.e. one cycle in 5 seconds, and if several meters are used it should be even lower. Unfortunately, at this frequency the component values for a typical demonstration turn out to be about 20,000 μ F and about 10,000H.

If one is prepared to allow the added complication of an additional d.c. polarizing voltage to be present in the circuit, the necessary capacitance can be obtained using electrolytic capacitors but an inductor of the required value is necessarily very large and expensive. Moreover, inductors as large as this tend to have very low Q values and saturate easily making some demonstrations, such as, for instance, logarithmic decrement of a ringing tuned circuit, impracticable.

A compromise is to simulate a large inductance with a "black box" utilizing some thermal device such as a thermistor or using other electronic circuitry. An example of this approach has been given in $W.W.^1$ In the writer's experience, however, students find such a black box unconvincing and the whole point of the exercise fails when it is necessary to resort to this technique, especially as in many cases the reactance of the device decreases with increasing frequency and it often needs a d.c. polarizing voltage before it will work at all.

Use of a strobing meter

A much more elegant approach is to use one of the new strobing meters which are now available and which do not require low frequencies or large components to allow complete investigation of an a.c. circuit. This type of instrument works by sampling any input waveform for a very short period once per cycle and integrating all of the sampled pulses to build up an exact copy of the waveform, but at a much lower frequency. The frequency of the sampling action is adjustable and by setting the sampling rate to be nearly that of the input frequency, a very low frequency representation of the input waveform is obtained in the same way as an optical stroboscope provides a slow motion representation of a periodic mechanical motion. The strobing meter provides a slow motion copy of an a.c. waveform on a centre-zero meter and the demonstration appears similar to the v.l.f. demonstration described previously.

The manner in which the instrument operates is illustrated in Fig. 1. As an example, a simple a.c. circuit containing resistance and inductance is shown and the waveforms of voltage and current are shown alongside.

A simple experiment used to demonstrate the phase of voltages developed across a resistance and inductance in series, in relation to the current phase.





The P.E.L. strobing multimeter which, as described in the text, can be used to investigate an a.c. circuit for demonstration.

The readings on the meters resulting from the sampling pulses are indicated, and these would be steady readings if the sampling rate was made equal to the supply frequency. In practice, for demonstration purposes, the sampling rate would be slightly different from the supply frequency and the sampling pulses which occur once per cycle would appear at a slightly different position in each successive cycle. The difference between the sampling rate and the supply frequency is the apparent display frequency and this can be as low as required. It would normally be adjusted to a convenient speed for the particular demonstration being given. The real frequency can be any suitable frequency available such as 50Hz or higher. Component values required are quite small and for a typical demonstration at 1kHz suitable values would be around $0.01 \mu F$ and 100mH. Several meters may be used together in this way and, providing the sampling pulses all occur together, each instrument indicates the voltage or current and its phase in the same way as the simple centre-zero meters indicate these values in the v.l.f. circuit described previously.

A good example of an instrument of this type is the Strobe Multimeter D40* shown in Fig. 2. This contains two independent multimeters in one case, permanently synchronized together so that relative phase as shown on the centre-zero meters is always correct. Each multimeter has eleven current ranges from $1.0 \,\mu$ A full scale to 500mA full scale, and ten voltage ranges from 0.5V full scale to 500V full scale. As the input resistance on the voltage ranges is $1M\Omega/V$ the instrument does not affect the working of the circuit to which it is connected.

The writer has found that this method of demonstrating the action of a.c. circuits is readily accepted by students as they are often familiar with the way in which a stroboscope works.

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* Manufactured by Physical & Electronic Laboratories Ltd., 28 Athenaeum Rd., Whetstone, London, N.20, of which the author is chief engineer.

Mass production of shadow-mask colour television tubes is well under way at the Mullard cathode-ray tube factory at Simonstone, between Preston and Burnley, Lancs. This plant, which started up in 1955 and has a production capacity of $1\frac{1}{2}$ million monochrome tubes a year, manufactures everything for the colour tubes except the shadow mask and the components for the electron guns. Covering 44 acres and employing 2,000 people, it has two distinct parts-a glass factory (incidentally one of the largest in the U.K.) and a tube assembly and screen laying plant. Assembly and screen-laying processes are basically similar to those outlined in a Wireless World article last year* but there is a greater proportion of automatic machinery than in the small plant operating. The production rate at Simonstone depends, of course, on the demand for colour television receivers, but Mullard say that by the end of 1968 the works will be capable of manufacturing 150,000 shadow-mask tubes a year. The two sizes of colour tubes, 25-inch and 19-inch, are made in alternate runs (of several weeks or months each) as parts of the machines have to be changed to handle the different sizes of components.

The glass factory, which produces the cones and faceplates, comprises three main parts: a weighing, dispensing and mixing area for raw materials; an oil fired furnace, operating at 1,550°C; and two automatic glass forming presses. The cone and faceplate of the colour tube require different types of glass, so the plant only produces one of these components at a time (one of the two presses acting as a standby). To change from one type of glass to the other takes about ten days. The furnace operates continuously and can handle 100 tons of glass a day.

What does all this mean to the customer? There does not seem to be any established retail price at which a shadow-mask tube can be bought over the counter. Last September Thorn-AEI Radio Valves and Tubes Ltd. said that the retail price for a 25-inch tube was "in the region of $\pounds 90-\pounds 100$ ". (Receiver manufacturers, however, can buy 25inch tubes for about $\pounds 55$ each and 19-inch tubes for about $\pounds 45$ each.)

*"Colour Tube Production", September 1967 issue.

Performance testing completed tubes. In this five-position test bay checks are made for blemishes in each colour and in white, for convergence, linearity, cathode quality and other characteristics. The work is speeded up by measuring aids such as digital read-out for colour purity tests, and the whole bay has a throughput of up to 30 tubes an hour.

Manufacturing Colour Tubes



Wireless World, June 1968

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Series Voltage Stabilization

A review of the properties of conventional voltage stabilizers and a practical circuit incorporating automatic overload protection without employing extra transistors

by Jan-Erik Sigdell*







Fig. 2 Illustrating the resistive path from input to output. The ripple in the base current modulates the collector current.



Fig. 3 Employing a current generator to eliminate resistive and capacitive paths from input to output.

Fig. 4 This circuit is bistable and tends to trigger into the cut-off state if the load is rapidly changed between two values.



Most transistor circuits for series voltage stabilization used today are of the type shown in Fig. 1-or of a similar type (the lower transistor is often replaced by a differential pair). This circuitry has the following two disadvantages. (1) There exist resistive or capacitive paths from input to output, which tend to raise the hum at the output and (2) there is no inherent protection against overload, either for the stabilizer or for the load itself. The first of these two disadvantages is especially serious when a path passes through an active component in such a way that its humraising influence is amplified (e.g. through the base-emitter diode of a transistor). This disadvantage is often overcome by more amplification in the feedback loop or e.g. by dividing resistor R, in Fig. 1, into two resistors and

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Fig. 5 A circuit that can be made to behave in a monostable fashion, the current cut-off state only being reached in the event of overload.





decoupling their interconnection with a capacitor or a Zener diode. The second disadvantage is often overcome with an extra circuit for current limiting or current cut-off.

Concerning the first disadvantage, it must be kept in mind that a configuration such as the one in Fig. 2 shall be regarded as having a resistive path from input to output (grounded side), as the ripple in the base current modulates the collector current. Concerning the second disadvantage the series transistor (transistors) should, at a determined current level, change over from being a part of a voltage stabilizer to being a part of a current stabilizer or a current cut-off circuit.

The first requirement can be fulfilled with the Zener diode decoupling method previously mentioned or by employing a current generator¹ as in Fig. 3 (if Zener diodes are regarded as ideal). Two other solutions that fulfil both requirements are shown in Figs. 4 and 5. Clearly stabilizers using these designs may become bistable and the only way to ensure monostability in the voltage stabilizing mode is to provide 'on' leakage currents. This is not unrealistic for the circuit of Fig. 5 as will be seen. On the other hand one may introduce extra current sources through reversely polarized diodes with sufficient leakage currents. Also the new field-effect current limiting diodes are of interest for this purpose, although as yet expensive. Another possibility is to put a zener diode across transistor Tr_3 , Figs. 4 and 5, which conducts when the output voltage is low but cuts off when it reaches its stabilized value. However, this requires a sufficient value of resistance in series with



Fig. 7 The current cut-off characteristic of the circuit of Fig. 6.

the Zener diode in order to limit the short circuit current at the output.

Bistability may also be an advantage, but only if the circuit automatically triggers itself into the voltage stabilizing state when the input voltage is applied and the output loading is not excessive. Furthermore, it should not trigger to the cut-off state if the load is suddenly changed between two permitted values. The circuit of Fig. 4 has both of these undesirable properties.

The circuit in Fig. 5, however, can be monostable, so that the cut-off state can be achieved only when an overload exists. This is due to the leakage current in transistor Tr_1 which is amplified twice in transistors Tr_2 and Tr_3 and drives the circuit from the cut-off state when the output loading is within limits.

A practical realization is shown in Fig. 6. The author has used this in a transistor communications receiver for almost a year. The diode D is included for temperature stabilization (it tends to cancel the change with temperature in the base-emitter voltage of Tr.). The capacitor C reduces the hum further. The resistor R raises the collector current of transistor Tr_1 to a more suitable working point (with a more constant input resistance). The circuit has the current cut-off characteristic shown sketched in Fig. 7. Its output resistance, measured statically from the change in output voltage at application of a 300 mA load, is 0.13Ω . The hum is less than 10 mV peak-topeak at a load of 300 mA.

REFERENCE

1: "Taschenbuch der Hochfrequenztechnik", Meinke/ Grundlach: p. 1626.

Circuit Ideas

A regular feature on original ideas in circuitry which have been found practical is to be started in *Wireless World*. Presented in the form of short notes, the items will be essentially functional "bricks" which somebody has found useful at some time. Performance, originality of realization and economy of components will be the most important criteria. Readers are invited to contribute to this series: the more ideas we get the better will be our selection.

Hanover Fair Grows Internationally

When considering exhibitions on a grand scale, the Hanover Fair ("Hannover Messe"), which began in 1947 as an exhibition of German manufactured products, must now surely rank among the biggest and most international in the world. It is a large self-contained entity situated in the rural outskirts of this West German city, with its own airport with light aircraft running a shuttle service to and from the main Hanover airport. An internal electric bus service gives visitors a free ride from one section to another and there are even travelling "pavements" from which visitors can admire the pleasant surroundings in which the exhibition is set, as they are glided slowly to their next point of call.

The Fair, held this year from April 27th to May 5th, comprises some 28 pavilions set in close proximity to each other as well as a large area of outdoor exhibitions, mostly of heavy machinery. Most of the pavilions would each be capable of absorbing the total area of, for instance, London's Earls Court pavilion, and it took three of these to house the electronics section alone. This section embraced electrical installation equipment, control gear, radio and television, aerials, business machines, computers and test and measuring equipment. In the television section the accent was, naturally, on colour, and a few of the more outstanding items of interest will be reviewed in the July issue of World. Wireless The majority of receivers demonstrating colour were operating on the PAL system but some were being shown which were intended for the French SECAM market. The Association of French Exporters of radio and TV sets had several stands showing SECAM colour pictures on 58cm receivers. They were using their own generating equipment which provided six channels of colour stills. On some of the screens a horizontal pattern could be seen but this was explained as being due to direct pick-up of interference from a nearby mains feed to the building. At any rate, even in this location SECAM could not suffer from the simple PAL defect—"Hanover blind"!

Some 5,500 exhibitors from Europe and overseas countries were at the Fair, which has come to be regarded as the shop window for world technology, and in passing it should be mentioned that products from Japan in particular were much in evidence. Sony were demonstrating a miniature television receiver measuring about $4 \times 7 \times 20$ cm long with a 2.5cm c.r.t. at one end. The circuit is designed around i.cs and it covers the full v.h.f. and u.h.f. bands. The receiver can operate from the mains or as a portable using a battery-pack. It costs in the U.S.A. about £100.



New Products

Low-cost Counter

A development of their 12.5-MHz frequency meter/counter/timer incorporating both i.cs and a memory store has been introduced by Venner at £245. Each counter board comprises three modules; a decade, a store and a display tube drive module. The instrument, type TSA 6636/2M, is designed for general purpose applications in industry and gives a 6-digit readout, including decimal point, with a figure size of 0.6in. An optional extra is a 1248b.c.d. printer output. Be-cause this output is available from the memory, the printer can operate while the next result is being measured, thus increasing the cycling rate over previous models. For frequency measurement (10Hz to 12.5MHz), gating times are 1 us to 10s in decade steps. Input sensitivity is 75mV into 250k Q. Single and multi-period measurement covers a 10Hz to 1MHz range with periods from 1 to 107. As a timer the TSA 6636/2M covers 1 us to 10s in decade steps, single and two-line start /stop facilities being provided. Provision is also made for



gated counting. Display time is variable from 0.5 to 5 seconds or infinite, and accuracy is ± 1 count \pm crystal stability ($\pm 1 \times 10^{-6}$). Venner Electronics Ltd., Kingston By-Pass, New Malden, Surrey.

WW 304 for further details

New Tracking Filters

Two new tracking filters which are based on the AIM voltage programmable filter (PFO 166) instead of the conventional heterodyne systems using mixers and oscillators, have been announced by AIM Electronics. Analogue computing techniques are employed to simulate a perfect LC filter in which the centre frequency may be varied by an external programming voltage. An application for tracking filters is in vibration studies where resonant frequencies are formed by applying vibrations of increasing frequency to an object under test. The tracking filter is required to eliminate noise and harmonic effects from the resonance detector output signal by following the



increasing vibration frequency. The System 5.18 tracking filter has a frequency range of 0.1Hz to 50kHz and is capable of sweeping at up to 50kHz over frequencies of 300:1 without changing switched ranges. System 5.19 is similar but includes an idependent swept oscillator which provides signals suitable for driving vibrators. AIM Electronics Ltd., 71 Fitzroy Street, Cambridge. WW 307 for further details

A.M./F.M. Signal Generator

Amplitude and frequency modulation facilities are combined in a new m.f./h.f. solid state signal generator TF 2002AS announced by Marconi Instruments. Based on the a.m. signal generator TF 2002, this new instrument incorporates several supplementary features which considerably extend its scope. TF 2002AS is suitable for all types of measurement on a.m. and f.m. receivers, i.f. amplifiers and demodulators within its frequency range-including dynamic measurements normally outside the scope of most signal generators below 30MHz. It covers the carrier frequency range 10kHz to 72MHz with provision for a.m. up to 100% at modulating frequencies from 20Hz to 20kHz. The internal modulating oscillator is tunable to any frequency in this range. An incremental tuning control is calibrated directly in frequency with symmetrical scales for positive and negative changes. Although provision has been retained for standardizing its accuracy against the internal crystal calibrator, the direct calibration remains valid within 15% for all main



tuning ranges above 100kHz without trimming. External frequency shift and remote carrier levelling facilities are both fully symmetrical. A d.c. control voltage variation of $\pm 1V$ produces frequency shift ranging from ± 0.5 kHz at 100kHz to ± 50 kHz on the highest frequency ranges. At the lower carrier frequencies, the shift range has been increased more than three times so that the swept-frequency bandwidth display facility has been usefully extended into the m.f. band. High control sensitivity permits utilization of the low voltage sawtooth ouput of transistor oscilloscopes. Carrier levelling, up to 1 ±100% variation on the initial output voltage, is obtained by application of up to ± 6 volts d.c. input. Price £987. Marconi Instruments Ltd., St. Albans, Hertfordshire.

WW 305 for further details

Linear I.Cs

Four new linear integrated circuits available from Mullard are enclosed in multi-lead TO-5 encapsulations. Operational amplifier TAA241 is designed for general purpose applications in instrumentation and control. It is a special version of the TAA243. Nominal supply voltages +12V and -6V, and operating temperature range is 0 to 7°C. Input and output impedances are $32k\Omega$ and 200Ω respectively. Offset voltage is 1.5 mVand offset current $0.5 \,\mu$ A. Large signal voltage gain is 3,400. The TAA300 is an a.f. amplifier which, with 10mW input will give an output of 1W. It is designed for 9-volt operation. Input impedance is 8Ω . Low-noise a.f. pre-amplifier TAA310 is designed for use in record/playback units. With a supply of 7V the voltage gain is greater than 95dB and input impedance exceeds 10k Ω . Operating temperature range is -20 to



+75°C. Amplifier TAA350 is a wideband, differential, limiting amplifier for use with an f.m. carrier at about 6MHz and is particularly suitable for television intercarrier sound amplifiers. Nominal supply voltage is 6V and input limiting voltage is 100 μ V. Input resistance is 2.5k Ω and output resistance 80 Ω . Typical voltage gain is 65dB. Mullard Ltd., Torrington Place, London, W.C.1.

WW 302 for further details

New Frequency Standard

Output signals in decades from 10kHz to 10MHz, with extremely close limits of frequency accuracy and stability, are provided by a new frequency standard, the FS3, from Advance Instruments. This instrument is particularly suitable as an external standard frequency where digital frequency meters and timer counters are required to measure to high accuracy, or for use as a precision laboratory standard to which other equipment may be referred, the aging rate being



as low as 2 parts in 109 per day after 30 days' operation. It incorporates an integral charger and rechargeable battery. Should the mains supply fail the instrument automatically continues to operate at full accuracy on the batteries for up to four hours. When the mains supply is reinstated, the battery is automatically recharged. The FS3 utilizes as its master oscillator an oven-controlled 5MHz crystal. The oven current is monitored for correct operation by a front panel meter. Extensive use of integrated circuits has been made in the design of this instrument. Price £325. Advance Electronics Ltd., Hainault, Essex. WW 303 for further details

Switching Programmer

A switching programmer measuring 8 \times 8cm in cross section and 25.5cm long, including the motor, is designed to replace 32 d.p.d.t. relays and related control devices which may be currently employed in conventional switching systems making it possible to greatly reduce wiring and space requirements. It also permits greater flexibility in



setting up and changing required switching patterns. Now available from Sealectro this model 96D programmer can control 96 s.p.d.t. contacts each of which is electrically isolated to control independent circuits. Contacts are rated to make and break 2A at 230V a.c. or 24V d.c. (resistive load) and the unit's memory drum can be programmed to actuate any number of contacts in groups of three at any of the 60 drum positions. Standard operating voltage is 230-240V a.c. Sealectro Ltd., Walton Road, Farlington, Portsmouth, Hampshire. WW 310 for further details

New T.T.L. I.Cs

Transitron Electronic, has announced three additions to its Series III range of t.t.l. integrated circuits. They are two 4-bit ripple counters and a 4-bit shift register. Both ripple counters are monolithic and have been designed for high speed, high noise immunity, high output capacitive and current drive capabilities. Both consist of four flip-flops which have been arranged to give a one-way binary ripple counter (TRC 2521-2524) or a binary code decimal ripple counter (TRC 2525-2528). Temperature range of both types is -55 to +125°C and other significant parameters include power consumption of 18mW at 5V; count frequency 25MHz; output current drive 20mA. Other features include low current loading, complete t.t.l. compatibility, and saturated "HLTTL" circuitry. Both counters find application in a wide range of multiplexing, data logging, and flight data recording situations. The new shift register (TSR 2511-2514) is also a monolithic 4-bit device, designed for high speed, high noise immunity, and a high output capacitive and current drive capabilities. Input and output from the four flip-flops are arranged in four different combinations to provide serial in-serial out, serial in-parallel out, parallel in-parallel out, parallel in-serial out operating modes. Enable /disable outputs do the mode selection and no external connections have to be made to switch from one mode to another. The new register can also be converted into left/right shift by simple external connections. Temperature range is -55 to +125°C; power consumption 195mW at 5V; shift frequency 25MHz; output drive 20mA; and power supply, single 5V. All three circuits are aveilable in either 14-lead flat pack, or 14-lead dual-in-line package. Transitron Electronic Ltd., Gardner Road, Maidenhead, Berkshire.

WW 301 for further details

Radio Communications Equipment

The first items in a new range of h.f. equipment introduced by Redifon Communications Division are the G450 1kW h.f. general purpose transmitter; the R499 10-channel m.f./h.f. s.s.b./d.s.b. receiver, and MCU6 aerial distribution amplifier. Particulars have also been given of the TT20 telegraph terminal unit which, although specially designed for the Royal Navy, is now available to other naval, military and commercial users. Type G450 is a self-contained s.s.b./d.s.b./c.w. transmitter without the need for external drive units. It operates on up to ten crystal-controlled channels in the 1.5 to 30Mhz frequency range with provision for an external frequency source such as a frequency synthesizer. The transmitter comprises a driver/p.a. unit, an exciter unit and a power unit, employing transistors in all but the drives and p.a. stages. The exciter unit can be used independently as a drive unit for other linear amplifiers. Type R499 is a fully transistorized receiver which is available in a number of versions to suit individual service and system requirements. Crystal filters are fitted internally according to the modes of signal to be received, and these are available for c.w., s.s.b., d.s.b. and i.s.b. The receiver operates on up to 10 channels in the h.f. band and, with the addition of filters, also on the 255-525 kHz band. Noise factor is generally better than 7dB and front end protection is provided against e.m.fs of up to 30V r.m.s. The a.g.c. system employed ensures that the gain of the i.f. stages is reduced before that of the r.f. stages,

and precautions are taken to prevent large signals from introducing intermodulation distortion. Remote control facilities (in conjunction with a remote control unit type RC116) include channel selection, service and on/off switching, fine tuning, a.f. monitoring and channel-in-use indication.

Type MCU6 aerial distribution amplifier has a dynamic range of 135dB extending from below $1\mu V$ to 5V total input e.m.f., over a working frequency of 95kHz to 30MHz. Using silicon planar transistors throughout, the circuit comprises a push-pull amplifier with an associated wide-band a.g.c. system controlling a front-end attenuator. The a.g.c. begins to operate at an aerial input e.m.f. of approximately 500mV and limits the output to about 500mV, thus protecting associated receivers and permitting continued reception of a wanted signal in the presence of strong signals which would normally block the system. Two versions of the MCU6 are available: type A accepts a single input and provides ten outputs; type B contains two independent amplifiers each with one input and five outputs. Constructed for standard 19-in. rack mounting with a $3\frac{1}{2}$ in. front panel, the MCU6 normally operates from



100-125V or 200-250V 45-65 Hz, or can be made to special order to operate from 24V or greater d.c. supplies, with positive or negative earth. Type TT20 telegraph terminal unit comprises

twin v.f. receivers for independent simultaneous operation, and common power supply unit. As part of a modern communications system, each receiver forms part of a single teleprinter channel and accepts either a two-tone frequency exchanged keyed, or a frequency shift keyed telegraph signal, which it converts to a 1kHz tone or to a low-level d.c. signal.

WW 316 for further details

I.C. Patchboard

A rapid means of interconnecting integrated circuits and discrete components is provided by a new patchboard system by Circuit Integration Ltd. All types of i.c. configurations are catered for and special carriers are available for mounting other electronic components. In use, the components are mounted on the carriers permitting them to be plugged-in to any one of the patchboard range. Power supply connections are automatically applied to each carrier, and interconnection between devices is carried out by means of plug-in links. Five sizes of patchboard are avail-



able, offering 12, 18, 24, 30 or 48 pin circuits. Circuit Integration Ltd., 99 Bancroft, Hitchin, Hertfordshire. WW 312 for further details

High-Frequency F.E.T.

A low-cost high-frequency n-channel field effect transistor suitable for r.f. mixers, amplifier front ends, switching and general purpose applications is available from Union Carbide UK Limited. The new f.e.t., type UC734, has a minimum mutual conductance of 3mA/V at 200MHz and a pinch-off voltage between -1 and -8V. The reverse transfer capacitance is 0.8pF. Price of the UC734, for quantities of 25-99, is 9s 8d each. Special selections are available to customer specifications for a nominal charge. Union Carbide U.K. Ltd., Electronics Division, 8 Grafton Street, London, W.1.

WW 308 for further details

Two-way Telephone Amplifier

Telephone conversations up to a distance of 25ft from the telephone are possible using the Magnafone Mark 10, say the makers of this two-way telephone amplifier. There is no electrical connection with the telephone, a hearing-aid type microphone picks up the speaker's voice which is amplified and conveyed via the telephone to the caller. The caller's voice is received through a loudspeaker. A number of persons can participate in the conversation. The instrument is batteryoperated and is contained in a case about the size of a telephone. In operation the handset is placed on top of the case. The microphone and loud-



speaker volume controls regulate between the maximum and medium gain, so that when they are set εt zero the instrument is still operative. Price of the Magnafone Mark 10 is 29gn. The Magnetic Broadcasting Co. Ltd., Paragon Works, Ruvigny Gardens, Lower Richmond Road, Putney, London, S.W.15.

WW309 for further details

Stabilized Power Supplies

Fully variable output with a choice of current ratings from 1 to 10A are the facilities offered by a new range of modular power supplies types PM16-19 announced by the Volstat division of Advance Electronics. They have been specifically designed to meet the requirements of integrated circuit technology, with particular reference to reliability, and they incorporate a new protection circuit to safeguard both the power supply and the load. Stability of 10,000 to 1 is claimed, and under the





worst conditions of maximum temperature output voltage and current, the estimated m.t.b.f. is given as not less than 30,000 hours. Advance Electronics Ltd., Hainault, Essex.

WW 318 for further details

I.C. Operational Amplifiers

Extremely high input impedance with very low offset current and input bias current are features claimed by Transitron for two new integrated circuit operational amplifiers, type TOA7709 (for the military temperature range -55° C to +125°C) and type TOA8709 (commercial 0°C to 70°C range). These devices are compatible with the TOA1709 and TOA2709, both in pin configuration and compensating circuitry. The exceptional input characteristics effectively eliminate bias current and offset current as an error source, and allow the use of high impedance levels in circuit designs. The TOA7709 features $3M\Omega$ minimum input impedance (typically $10M\Omega$), input bias current of 10nA maximum, and the TOA8709 a minimum input impedance of $1M\Omega$ (3M Ω typical); input bias current of 60nA and input offset current of 20nA maximum. Both are available in dual in-line packs, flat packs and TO99 cans. Transitron Electronic Ltd, Gardner Road, Maidenhead, Berks.

WW 342 for further details

Polycarbonate Film Capacitors

Mullard has recently added a range of polycarbonate film capacitors to its 344 series. The new additions are 100V type capacitors with pin connections designed for direct insertion into the 0.1in grid of a printed circuit board. The regular rectangular outline of the capacitors permits high packing densities. With capacitance values of 0.68 to 6.8μ F, the new range complements the 0.01 to $2.2\mu F$ coverage provided by the popular 250V range. Because of their lower working voltage the new types are physically smaller; for example a $0.1 \mu F$ in the 100V range is the same size as a 250V 0.01μ F. Special features of the 100V range are: low loss (tan & $\approx 30 \times 10^{-4}$) and long term stability with a d.c. load (change of capacitance is less than 1.5% at 2.5°C). Mullard Ltd, Mullard House, Torrington Place, London, W.C.1. WW 343 for further details



Integrating Millivoltmeter

Suitable for both instantaneous and time-integrated (averaged) voltage measurement, a dual purpose millivoltmeter by Time Electronics, a new company, uses an f.e.t. chopper stabilized amplifier operated in a negative feedback mode. An input resistance of $1M\Omega$ is maintained throughout the entire range of 1mV to 30V. The integrator section employs the amplifier in an error-correcting feedback integrator circuit and the time integral is displayed on a 180°, 5" mirror scale meter, zero to f.s.d. representing 36mV/sec (1/100mV/hr) for the mV/hr range, and 36V/sec (1/100 V/hr) for the V/hr range. A 6-digit magnetic counter is used to total these units of integral by advancing one count each time the meter reaches f.s.d. The mean value of the input voltage over a given period is then determined by dividing the number of counts during the period by $100 \times \text{period in hours}$, the result being in mV or V depending on the selected range. The integrator output and count



command are available at the rear of the instrument for driving external equipment. Dimensions $20 \times 13 \times 13$ cm; price £98. Time Electronics, 10B High Street, Swanley, Kent. WW344 for further details

Video Tape Splicer

Increased use of video tape recorders using $\frac{1}{2}$ in. magnetic tape has led to the introduction of a new $\frac{1}{2}$ in. splicer ty Multicore Solders, based on the well-known Bib splicer for $\frac{1}{4}$ in. tapes. Model 21 splicer enables diagonal or butt joints to be made. The recommended retail price of the splicer which



is packed in a kit with six razor cutters, a reel of splicing tape and a Bib size E tape head maintenance kit, is £9 10s. Multicore Solders Ltd., Maylands Avenue, Hemel Hempstead, Herts. WW 311 for further details

Linear Microcircuit

A fully integrated low-cost differential amplifier. is the latest linear microcircuit, the μ A730, from SGS-Fairchild. It is designed to replace two discrete dual-transistors as a gain block in general purpose a.c. and d.c. applications. Encased in a low profile 8-lead TO-5 can, the new device has an input offset voltage of 2.5mV maximum and an inputoffset current of 500nA. Common mode rejection ratio is 70dB minimum, differential voltage gain is 145. Any combination of single-ended or differential configurations can be employed at its input and output, making the μ A730 suitable for use as a voltage comparator, a phase splitter, level detector, voltage regulator or a push-pull amplifier. Other uses include preamplifiers for sensing; magnetic tape, audio summing and instrument amplifiers; voltmeters and transducers. SGS-Fairchild Ltd., Planar House, Walton Street, Aylesbury, Bucks.

WW 345 for further details

High-frequency Analogue Multiplier

Quarter-square Circuit Operates at Megahertz

by M. E. Whatton, A.M.I.E.E., and G. Crisp

D URING an investigation into a colour television system^{1/2} it was necessary, as part of the process of decoding, to devise a method of multiplying together the two received colour subcarrier signals. A known technique for achieving this is the use of a balanced or ring modulator, and successful multipliers were built using this technique.

Circuits for multiplication are, however, well known in the analogue computer field, and it was thought that an investigation into the possibilities of using such circuits would prove useful. This article shows the techniques used and describes a suitable multiplier. The colour subcarrier frequency was 4.43MHz and the unit constructed had a bandwidth of about ± 1.5 MHz. The multiplier was based on the "quarter-square" multiplier which has been successfully used in analogue computer techniques³.

The reason for the name "quarter-square" multiplier is as follows. If it is desired to multiply together two voltages V_1 and V_2 , then this may be carried out as shown below:—

 $(V_1 + V_2)^2 = V_1^2 + 2V_1V_2 + V_2^2$ $(V_1 - V_2)^2 = V_1^2 - 2V_1V_2 + V_2^2$ $\therefore (V_1 + V_2)^2 - (V_1 - V_2)^2 = 4V_1V_2 \dots (1)$

Thus to obtain the required term V_1V_2 , it is necessary to perform the squaring and other operations on V_1 and V_2 as shown on the left-hand side of equation (1) and then divide the resultant answer by four. Hence the term "quarter-square".

The component parts of the multiplier are therefore:

(a) An adder to produce $(V_1 + V_2)$ and $(V_1 - V_2)$

- (b) \vec{T} wo square-law devices
- (c) A difference amplifier

A block schematic of the multiplier is shown in Fig. 1. As the particular square-law device used requires "push-pull" input signals these are also provided by the adding network.

M. E. Whatton joined the B.B.C. operations and maintenance department from Cinema-Television Ltd. in 1947 after qualifying at the South East London Technical College. In 1952 he transferred to the designs department and is now in the television studio section.

Graham Crisp, who is 23, joined the B.B.C. in 1963 as a technical assistant in the transmitter operations and maintenance department. He went to the designs department in 1965 and is now in the television measurements section.





Fig. 2. Circuit for obtaining square-law function of input, Vin.

The Square-law Unit

A simple square-law device⁴ was developed which consisted of two transistors feeding a common collector load as shown in Fig. 2.

The collector current/base voltage curve of either transistor may be represented by the following equation:

 $I_{c} = a + bV_{be} + cV_{be}^{2} + dV_{be}^{3}$ where a, b, c and d are constants. Therefore $I_{c_{1}} = a + bV_{in} + cV_{in}^{2} + dV_{in}^{3}$ $I_{c_{2}} = a - bV_{in} + cV_{in}^{2} - dV_{in}^{3}$ Then the total current

 $I_{\rm cc} = 2a + 2cV_{\rm in}^2$

The first term is a constant and the remaining term is proportional to the square of the input voltage. Therefore the change in output voltage is proportional to the square of the change in input voltage.

Circuit description

The circuit of the multiplier is shown in Fig. 5. The two high-frequency signals V_1 and V_2 are applied to two broadband r.f. transformers T_1 and T_2 . These transformers have a centre-tapped secondary winding which is bifilar wound. Transformer T_1 produces two signals V_1 and $-V_1$ across its secondary winding and transformer T_2

Fig. 1. Schematic showing principle of operation of the quarter-square multiplier.

produces V_2 and $-V_2$ across its output. These signals are combined in simple resistive adders R_1 to R_8 to produce the four signals, $(V_1 + V_2)$, $-(V_1 + V_2)$, $(V_1 - V_2)$ and $-(V_1 - V_2)$. These four signals are phased by the split-stator capacitors C_5 and C_6 .

 $V_1 \times V_2$

As the square-law transistors are voltage operated it is necessary for these to be driven from low impedance sources. These are provided by the emitter followers Tr_1 , Tr_4 , Tr_5 and Tr_8 . The square-law transistors Tr_2 and Tr_3 are balanced by the network R_{13} , R_{15} and R_{16} . In a similar way transistors Tr_6 and Tr_7 are balanced by the network R_{25} , R_{27} and R_{28} .

 R_{25} , R_{27} and R_{28} . The signal at the collectors of Tr_2 , Tr_3 is then $(V_1 + V_2)^2$ and at the collectors of Tr_6 , Tr_7 is $(V_1 - V_2)^2$. These two signals are applied through emitter followers Tr_9 , Tr_{10} to a simple difference amplifier Tr_{11} .



Fig. 3 Frequency response of the multiplier.



Fig. 4. Input/output characteristic of the multiplier.

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The common mode rejection of this amplifier is adjusted at low frequencies by R_{40} and R_{38} and at high frequencies by C_9 and C_{10} .

The output signal from Tr_{11} consists of two parts: the first is the required output signal and the second is a high frequency component at approximately twice input frequency. This high frequency component is removed from the output by the networks R_{44} , C_{11} , L_2 and the network R_{46} , C_{12} . The signal is then fed to the output through the emitter follower Tr_{12} .

Fig. 3 shows the frequency response of the unit, and this is about 1.5 dB down at 1MHz. Fig. 4 shows the input/output characteristics for V_1 and V_2 . This is linear over a range of 23dB for each input. Fig. 6 shows the output waveforms for output frequencies of 200kHz and 1MHz both above and below 4.43MHz.

Fig. 6. Output waveforms for output frequencies 200kHz and 1MHz above and below 4.43MHz.



Conclusion

The investigation showed that a quartersquare multiplier could be built and used at colour sub-carrier frequencies. In such a multiplier, the input/output characteristics were very linear for a wide range of inputs and the frequency response adequate.

In the circuit as described no attempt has been made to design an economical unit and the circuit was built only to investigate the techniques required. As over half of the transistors were used as emitter followers it seems that some of these could be eliminated in any practical application.

The authors would like to thank the Director of Engineering of the B.B.C. for permission to publish this article and also many of their colleagues in Designs Department for their assistance.

References

- 1. The NIR system. See "From Russia with Love", Wireless World, February 1966, p. 73.
- *ibid.* "B.B.C. Man Anticipates Russian Colour TV System?", p. 75.
 "Waveforms" by Chance, Hughes, MacNichol,
- Waverorms' by Chance, Hugnes, MacNichol, Sayre and Williams. McGraw-Hill Book Company, New York (1949), Section 19.2, p. 668.
- 4. *ibid.* Section 19.7, p. 683.

Wireless World, June 1968

World of Amateur Radio

R.A.F. Golden Jubilee Year

A celebration dinner organized by the R.A.F. Amateur Radio Society to mark the 50th Anniversary Year of the Royal Air Force will be held at No. 1 Radio School, R.A.F. Locking, near Weston-super-Mare, on Saturday, July 6th. Activities will commence at 13.30 hours. Provisional plans for the day include a display of historical wireless equipment and visits to the Headquarters station and to selected training laboratories at the Radio School. There will also be a mobile frequency measuring contest and a mobile field strength contest. Talk-in stations will operate on 160 and 2 metres. Entertainment will be provided for the ladies and children.

The celebration dinner will be held in the Grand Atlantic Hotel, Weston-super-Mare. Dress will be informal and the charge for the dinner will be \pounds 1.11.6. Applications should be sent to reach the Treasurer, R.A.F.A.R.S., Royal Air Force, Locking, Weston-super-Mare, Somerset, by mid-June.

England-France on 13cm

For the first time, on February 18th, French and British radio amateurs succeeded in establishing communication across the English Channel on 13 cm (2350 MHz). The distance covered was 35 km the stations being located at Cap Blanc-Nez (15 km west of Calais) and on cliffs near Dover Castle. At the French station (F2F0/P) a 2C39 was pulse-modulated at 200 Hz delivering 1kW peak output at 2350 MHz to a one-foot dish 400 feet above sea level. The English station (G3RPE/P) used two transmitters. One had a DET22 self-excited oscillator with transistor modulator (input 8 watts and A3 output of 300mW at 2320MHz). The other had a self-excited oscillator/amplifier using specially-made experimental transistors. The output was chopped 1kHz, the input was 2.5 watts and the mean output power was 400mW. On the receiving side a crystalcontrolled convertor (2350-2352 MHz) fed a narrow-band (4 kHz) tunable transistor i.f. strip. The aerial was a 4-ft dish, mounted 8ft above ground on a site about 350ft above sea level.

This amateur "first" came 37 years after the first public demonstration of what was then called "micro-ray" communication between Dover and Calais on 17cm and 35 years after a commercial radio link on 17cm was established between Lympne, Kent, and St. Inglevert, France.

International Mobile Meeting. The annual international meeting organized by the Amateur Radio Mobile Society will take place on Sunday, June 30th, at the United States Air Force Base, R.A.F. Mildenhall, Suffolk. Camping and caravanning will be permitted from the Friday evening. The Trade Show (a feature of A.R.M.S. Mobile Meetings) is to be accommodated in an aircraft hangar and the organizer (F. J. Barns, G3AGP, 60 Alverstone Avenue, New Barnet, Herts) will be pleased to receive enquiries from firms or organizations who wish to participate. The secretary of A.R.M.S. is Norman Fitch, G3FPK, 79 Murchison Road, Levton, London, E.10.

Echo of the Q.M. At an informal ceremony at the U.S. Embassy in London, Mr. Eric Godsmark of the Radio and Broadcasting Department of the Post Office recently received a certificate and plaque from the radio amateurs of Long Beach, California in appreciation of the services he rendered to those U.S, amateurs who were authorized by the U.K. Postmaster-General to operate a station on the Queen Mary during her last transatlantic voyage. Mr. Godsmark organized arrangements for a special amateur licence and call-sign (GB5QM/MM) which enabled American radio amateurs from California to contact, from the Q.M., other licensed amateurs throughout the world.

Morse Proficiency Transmissions. The Royal Naval Amateur Radio Society located at H.M.S. *Mercury*, Leydene, Petersfield, Hampshire, transmits morse code proficiency runs on 3520 kHz at 19.00 G.M.T. on the first Tuesday in each month at speeds of 20, 25, 30, 35 and 40 words per minute. A proficiency certificate is issued for 100% copy of a particular speed. The transmissions are made from the Headquarters' station G3BZU which also operates daily on 7010 kHz \pm 5kHz between 11.00 and 12.00 G.M.T. Claims for code proficiency certificates should be sent to the above address enclosing an I.R.C.

Swedish V.H.F. Beacon. A new v.h.f. beacon is now operating on 145.960 MHz from Spansberget, near Borlange, in Central Sweden. The main purpose of the beacon is to produce a signal for the aurora back-scatter investigations which are being carried out at the Max-Planck Institute, Lindau,

Germany, by Dr. G. Lange-Hesse, DJ2BC. The establishment of the beacon station was made possible by close co-operation between the Swedish Telecommunication Authorities and the Swedish national amateur radio society (S.S.A.). The call-sign of the beacon is SM4MPI transmitted once a minute followed by a 50-second dash. The aerial consists of four 5-element arrays and has a gain of about 12dB. The beam is directed through 325°. Power output is around 100 watts and although the array is only 12 metres (40 ft) high the base of the mast is almost 500 metres (1650 ft) above sea level. The exact location of the beacon is N 60° 22', 58", E 15° 08' 30". Reports on signals from the beacon will be appreciated by G. Eriksson, Box 12, Falun 1, or by the S.S.A., Fack, Enskede 7, Sweden. One of the most important items of information to be included in a report is whether or not the signal was received via aurora back-scatter or by direct path.

U.H.F. Call-sign Directory. The first edition of a new directory of amateur radio stations active on 220 MHz and higher frequency bands is now available from the headquarters of the American Radio Relay League, 225 Main Street, Newington, Connecticut, 06111, U.S.A. The directory lists the name and address of each operator together with information about his interests in such diverse fields as meteor scatter, moonbounce and OSCAR reporting. Information is also given on aerial systems in use. A copy of the directory may be obtained from the above address by sending a large self-addressed envelope and four I.R.Cs.

German Training Courses. The outstanding success of the intensive training courses for prospective radio amateurs organized last year by the German national amateur radio society (D.A.R.C.) has prompted that organization to offer a more extensive programme for 1968 and at the same time to invite participation from foreign students who can speak or read German. The following is a list of the dates of this year's summer holiday courses: July 21-August 7 South Bavaria. July 25-August 14 Wurtemberg.

July 28-August 13 Lower Saxony.

August 1-21 Silesia-Holstein.

August 4–21 Hessia.

Further details of the training courses can be obtained from H. J. Henske (DL1JH), 605 Offenbach/Main, Wiesenstrasse 21.

Knokke Convention. The 4th International Amateur Radio Convention organized by a group of Belgium Channel-coast amateurs is to be held in the Casino, Knokke, during the weekend September 13th-15th. Details of the programme can be obtained from Lucien Vervarcke (ON4LV), Lippenslaan 284, Knokke 1, Belgium.

R.N.A.R.S. During the Lee-on-Solent Air Day on June 15th, the Royal Naval Amateur Radio Society will be providing talk-in facilities on 1910kHz, 70.26MHz and 145.3MHz. Callsign GB3RN.

JOHN CLARRICOATS G6CL

Questions on page 161



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SWITCHED ILLUMINATED INDICATORS Devt. No. 1037

An attractive legended button unit of moulded construction lit by two low voltage L.E.S. lamps in combination with mains rated switching which can be either D.P.C.O. Push/Push (Successional) action or from 1 to 4 pole Make and Break or Change Over biased action. Legending to customers own requirements which can be carried out over the full Filter area or split into two independently lit sections. left and right, with light separation barrier between.

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RANGE OF MOULDED SWITCHES List No. S.270/2 (Toggle operated)

A new range of D.P.C.O. panel mounting Switches of moulded construction with Quick Make & Break action and self-cleaning silver contacts. Equally suitable for 'mains' or 'low voltage' uses. The range which is to be extended later this year will include various forms of actuation, i.e. Toggle, Biased Toggle, Push Button, Push On/Push Off (Successional Action), Semi-Rotary, & Key.

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A pair of Battery Holders accepting one (List No. B.16) or two (List No. B.17) U.2 cells. Both models have polished black front of panel parts with the front cap legended BATTERY which fixes by a push and twist action with Key Slot removal. Four 6BA. bolts fix the units to the panel and the cable connection is by tags accepting solder or 187 series push-on connectors. This range is to be extended later this year enquiries invited.



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WW-103 FOR FURTHER DETAILS

1. (c). Both junctions may become forward biased when a transistor is used in a switching circuit.

2. (c). Only carriers injected from emitter into base take part in transistor action. Hence the emitter current must consist as largely as possible of carriers moving in that direction. In a forward biased p-n junction, currents carried by holes moving from p to n and electrons from n to p depend upon the relative doping of the two sides. The carriers when injected into the base become minority carriers. The relatively large equilibrium population of minority carriers in the base (because of low doping) has no effect on the action.

3. (c). The transit time of the injected carriers from emitter to collector must be short compared to the minority carrier lifetime in the base material.

4. (c). α is the product of "injection ratio", "base transport factor" and "collector multiplication factor".

5. (c). The collector-base junction depletion layer increases in width, thus reducing the distance through which the carriers must move before they are swept into the collector.

6. (b). The purpose is to make the transit time of carriers in the base short, to produce a device with good high frequency performance.

7. (b). The diffusion method of doping must leave doping gradients in the material. Silicon planar transistors of p-n-p type are made; n-p-n are more common. A silicon planar transistor cannot be symmetrical.

8. (a) The net charge in the base must remain zero, so that extra electrons will be drawn in through the base lead, when the current starts, to compensate for the extra holes injected.

9. (c). If the junction were forward biased a current would flow between gate and channel and the input impedance would be lowered.

10. (d). In normal use the polarity of the drain voltage is such that the gate-channel voltage is greatest near the drain. Hence at this point the depletion layer is wides t and the channel narrowest.

11. (c). It extends completely across the channel at the drain end. If the drain voltage is increased above the value at which this occurs an increasing length of the channel is blocked in this way.

12. (b). In an enchancement m.o.s.t. the channel is formed by applying a voltage to the gate that attracts minority carriers to the surface under the insulation and thus induces a channel of the opposite type to that of the bulk material.

13. (b).

14. (a). A depletion m.o.s.t. has very similar properties to those of a junction f.e.t. Its input impedance, however, is higher.

15. (d). In regard to factors (a), (b) and (c), m.o.s.ts are no worse than other transistors.

16. (b). In regard to factors (a), (c) and (d), f.e.ts are at least no better than bipolar transistors.

Literature Received

Second editions of the numbers one and two handbooks, "Services for British Exporters" and "ECGD Credit Insurance and Financial Support Services" are now available from any Board of Trade regional office or the U.K. Publicity Section, Information Division, Board of Trade, 1 Victoria St., London, S.W.1. "Services for British Exporters" presents comprehensive information on the whole range of Government and other export services and explains, in 180 pages, how exporters can best benefit from them. The "ECGD Credit Insurance and Financial Support Services" handbook describes the facilities provided by the Exports Credit Guarantee Department. WW401 for further details

Display Devices. The display department of Ferranti Ltd, Gem Mill, Chadderton, Oldham, Lancs., have produced an eighteen-page brochure describing some of their products. Included are the essential electrical characteristics of the Microspot range of cathode-ray tubes with suggested applications, the types, and properties of the various phosphors available are also outlined. Details of other devices from custom built display equipments, including coils, flash tubes, trigger tubes (arc discharge and thyratron) to pulse transformers are included.

WW402 for further details

Now available from SASCO, P.O. Box 20, Gatwick Rd., Crawley, Sussex, the latest edition of their components catalogue. WW403 for further details

"Don't Trip Up" is the title of a leaflet describing rubber conduit strip which is designed to protect loose cables and wires, and staff from injury. Vulcascot Ltd., Acorn House, Victoria Road, London W.3. WW415 for further details

An adaptor that converts a standard b.n.c. input terminal on test equipment to a twin 4mm outlet, and also cylindrical coaxial terminated metal cases for packaging test circuits (voltage dividers, attenuators etc.) are described in a leaflet from Radiall Microwave Components Ltd., Station Approach, Grove Park Road, Chiswick, London W.4.

WW416 for further details

"What to Look for When You Buy a Car Radio" is the rather misleading title of a new booklet from Philips. It does not give technical advice to those about to choose a car radio as the title suggests, but describes the range of Philips car radios and gives details of which accessory kit is required for particular types of car, together with a full price list. Philips Electrical Ltd., Century House, Shaftesbury Avenue, London W.C.2.

WW417 for further details

The "Abridged Valve Data Book" summarizes the range of valves available from the English Electric Valve Company, Chelmsford, Essex. The book incorporates a thumb indexing system and is divided into three major groups. These are power, microwave and light conversion applications, with a final subsidiary group listing such products as lasers, cold-cathode tubes and vacuum capacitors.

WW418 for further details

"A Comprehensive Catalog and Guide to Operational Amplifiers" contains 36 pages of advice, data and product information relating to equipment manufactured by Analog Devices, Inc., 221 Fifth St., Cambridge, Mass. 02142, U.S.A.

WW419 for further information

Rather belatedly we bring readers' attention to the new components catalogue produced by Home Radio Ltd., 187 London Road, Mitcham, Surrey, costing 7s 6d plus 2s postage. The catalogue has 256 pages and lists over 7,000 items, 1,300 of them being illustrated. As is their usual practice the catalogue includes five vouchers worth 1s each.

BS 3363:1968, which is concerned with letter symbols for semiconductor devices, is now available and can be obtained from the British Standards Institution, 2 Park St., London W.1, price 12s.

We have received a copy of the British Amateur Electronics Club Newsletter. This describes in some detail the logic of a noughts and crosses computer being produced as a B.A.E.C. project. Among the other items incorporated is a description of a unit capable of demonstrating the operation of a colour TV tube. The chairman of the club and editor of the newsletter is Mr. C. Bogod, "Diebeus", 26 Forest Road, Penarth, Glam. WW413 for further details

Waveguide switches are described in a 15-page brochure available from Silvers Lab., Stockholm 42, Sweden. WW407 for further details

"A Pocket Guide to Semiconductors" is the title of a pocket book produced by Quarndon Electronics (Semiconductors) Ltd., Slack Lane, Derby, which gives prices and data on a variety of transistors, diodes and integrated circuits available from them. Quarndon are distributors for Texas Instruments, SGS-Fairchild, Transistor A.G., Emihus, Tadican, and Sprague.

WW404 for further details

A 15A a.c. constant voltage regulator with an accuracy of 0.3% and introducing only 3.5% distortion is described in a leaflet from Claude Lyons Ltd., Valley Works, Hoddesdon, Herts. The regulator, weighing 18lb, has a time constant of 0.1sec and does not employ any form of filtering to achieve the low distortion level.

WW405 for further details

H.F. Predictions—June

The curves, drawn by Cable & Wireless Ltd., show median standard MUF, optimum traffic frequency (FOT) and lowest usable frequency (LUF), the latter being for reception in the U.K. of medium-power commercial telegraphy transmitters using high-gain aerials. The LUFs by their relative proximity to FOTs, are a guide for all types of service.

Predictions are based on an Ionospheric Index (IF2) of 134 which is expected to be the current cycle maximum. Ionospheric and magnetic disturbances of the type which can last for several days reach peak intensity in the few years following sunspot maximum. As such disturbances have been at low level for some time serious interruptions are not expected.



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WW-108 FOR FURTHER DETAILS

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ELEMENTS AND CIRCUITS FOR AUTOMATIC CONTROL

By T. PUCHALKA and A. WOZNIAK.

English translation edited by N. G. Meadows, Ph.D., B.Sc., C. Eng., M.I.E.E.

This book gives examples of typical elements of control systems and discusses their applications in systems for automatic control.

It opens with a chapter dealing with fundamentals of control theory, which is of interest in its own right and also gives a fundamental basis for subsequent work. This is followed by an extended treatment of the dynamic behaviour of electrical circuits and electrical machines as elements in control loops, together with various aspects of non-linear behaviour. The third chapter is devoted to a discussion of selected elements of electrical control systems, classified on the basis of single-loop block diagrams. Finally, the book deals with stability criteria, stability synthesis and the general dynamical analysis of selected systems.

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V.H.F. SIGNAL GENERATOR MARCONI TF-801A/I. Covers 10 to 310 Mc/s. (4 bands). DIRECTLY calibrated. Int. Mod. at 400, 1,000 and 5,000 c/s. Attenuated or force output. Guaranteed overhauled, r force output. Guaranteed overhauled ccurate and in perfect working order 35. Carr. £1.

BEAT FREQUENCY OSCILLATORS. MARCONI TF-195M. Covers 10 cps. to 40 kc/s. in two sweeps. 0 to 20 kc/s. and 20 to 40 kc/s. Output 2 watts into 600 or 2,500 ohms. Panel meter indicates output voltage. A.C. mains operation 100 to 250 volts. First class condition. Fully tested. £20. Carr. 30/-.

AMERICAN HEADSET TYPE HS-30-U 600 impedance. BRAND NEW and boxed, 15/-. postage 2/6.

DISTORTION FACTORMETER MARCONI TF-142E. This instrument measures the percentage of total harmonic distortion in the fundamental frequency range 100 to 8,000 c/s. The lowest scale engraving is 0.05%, Will handle 2 watts (continuous) and willgive satisfactory readings with only I mW input. Mains operated. Output impedance 600 ohms. Very good condition. £29. Carr. 20/-.

MICROAMMETERS R.C.A. 0-500 microamps. 2±in. circular flush panel mounting. Dials are engraved 0-15, 0-600 volts. As used in the American version of the No. 19 set. BRAND NEW and boxed 15/-. P. & P. 1/6.

AR-88 SPARES AR-88 SPARES Knobs, Medium size, Set of 8 Knobs, Large size Condenser (3 ×4 mfd.). Post 4/6 Mains Trans. (L.F.) (postage 9/-) Escutcheons (Windows) 10/-5/6 5/0 12/6 42/6 8/6 MINIATURE RELAYS 240 v. A.C. coils. Contact assembly "makes" and I C.O. 5 amps. Size $2 \times |7\frac{1}{2} \times$ lin. Unused and removed from brand new equipment 8/6 post paid. MOVING COIL PHONES. Finest quality Canadian with chamois ear-muffs and leather-covered headband. Noise exclud-ing and supremely comfortable. Complete with moving coil microphone 25/-. DLR-5 Low impedance headphones with attached throat microphone. 12/6. All these items BRAND NEW. Postage extra 2/6. CINTEL NUCLEONIC SCALERS

. 36402 and 36411. Unused with hand-k. List Price £300/£320. Our Price £65. book

PACKARD-BELL PRE-AMPLIFIER Fitted with 6SL7GT and 28D7 Valves. Brand new and boxed with manual. 12/6. Postage 4/6. Brand

CRT Type 89D as used in the Cossor 1035 Oscilloscope. Brand New 59/6. P. & P. 4/6.

H1B Audio Signal Generator	£30	0
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J2B Audio Signal Generator	£35	0
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VM77C AC Millivoltmeter	£40	0
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These are current production, manufac	tured	in
U.K. by Advance Electronics Ltd. (not	disco)n-
tinued models). Showing a saving of	appro)X-
imately 33 $rac{1}{3}\%$ on nett trade price. BRAN	DNE	W,
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COSSOR OSCILLOSCOPE TYPE XT476 Detailed specification sent upon request. Offered in first class con-dition at £350. List price approximately £800. Detailed

WIRELESS SET No. 76 A compact CW only crystal controlled transmitter. Consists of a Pierce crystal oscillator (807) and a Power Amplifier (807). Both are cathode keyed by means of a relay. Six switched crystal channels are available in the frequency range of 2 to 12 Mc/s. (Crystals not included.) Aerial current is indicated on a panel meter and two spare valves are supplied. Operates from 12 v. car battery via internal rotary transformer. RF output 9 watts. Contained in steel case 12x 12×8in. Weight 30 lbs. Ideal for 80 or 40 meters or cheap enough for breakdown. Condition as new. Circuit included. £4/5/-. Carr. 10/-.

PRICES NOW REDUCED CINTEL EQUIPMENT. ELECTROLYTIC CAPACITANCE AND INCREMENTAL INDUCTANCE BRIDGE No. 36601 A modern instrument, all solid state, which accurately measures the capacity of electrolytic condensers from 0.1 µF to 1,000µF under operating conditions. Leakage current and polarizing voltage are separately metered. Inductances from 100 mH to 100 H can also be measured with current up to 100 mA. A.C. mains operation. Unused with handbook. List price £220. Our Price £70. WIDE RANGE CAPACITANCE BRIDGE. No. 1864. A matching instrument to the above. All solid state. Mains operation. Measures from 0.002pF to 100µF. Unused with hand-book. List Price £250. Our Price £75.

MARCONI TEST EQUIPMENT PORTABLE FREQUENCY METER TYPE TF.1026 SERIES TF.1026/4 2,000/4,000 Mc/s., TF.1026/5 1,800/2,200 Mc/s., TF.1026/6 3,800/4,200 Mc/s., TF.1026/7 1,700/2,100 Mc/s., TF.1026/6 3,800/4,200 Mc/s., 1,700/2,100 Mc/s., TF.1026/6 3,800/4,200 Mc/s., 1,700/2,100 Mc/s., TF.1026/9 2,425/2,525 Mc/s., 40 each. WIDE BAND MILLIVOLTMETER TYPE TF.1371 100µv to 300 mv in five ranges. 30 c/s. to 300 mc/s. 100µv to 300 mv in five ranges. 30 c/s. to 300 Mc/s. D.C. measurement 0.15 to 100 v., 20 c/s. to 300 Mc/s. D.C. measurement 0.15 to 100 v., 20 c/s. to 100 Mc/s. Can also obm, 50Ω to 5mΩ in 2 ranges. 10 c/s. to 10 Mc/s. Can also be used as a wide-band amplifier. 450. DAAY GENERATOR TYPE TF.1415. Provides sweep-delaying facilities when used in conjunction rist the TF.1330 (Series) or similar oscilloscope. Alternatively, transcilloscope. Alternatively, timay be used independently as a general purpose delay generator. 23. TE867.A Standard Signal Construct Test **MARCONI TEST EQUIPMENT** £35. TF.867.A Standard Signal Generator. £200 TF.874A Moisture Meter. £20 TF.1066.B/2 U.H.F. F.M. Signal Generator. £200 TF.1066.B/2 U.H.F. F.M. Signal Generator. £200 TF.1066.B/2 U.H.F. F.M. Signal Generator. £40 TF.102. Amplitude Modulator. £40 TF.121. Heterodyne Unit. £125 TF.122. U.H.F. Signal Generator (cabinet soiled) £125 TF.1262. U.H.F. Millivoltmeter £20 TF.1264. Slotted Line £20 TF.1275. V.H.F. Bridge Oscillator. £40 TF.1343. X-Band Signal Generator Set £125 TF.1350/1 Power Unit for TF.1346/1. £40 TM 5683 Attenuator £10 £10 £10 £25 TM 6156 Attenuator.... TM 6629 A.F. Signal Compressor Detailed technical specifications supplied upon request. Offered BRAND NEW at fraction of original cost. Carriage and Postal Charges to N. Ireland and Eire extra. CHARLES BRITAIN (Radio) LTD. II UPPER SAINT MARTIN'S LANE

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AR.88 VIBRATOR POWER SUPPLY UNIT. Operates from 6-8 volt D.C. supply. Output 300 volts, 90 ma. Brand new, boxed, complete with leads. 15/-, postage 7/6.

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closed-circuit system made by Britain's largest manufacturer of electronic equipment. The basic system comprises two units-camera and control monitor. The units are fully transistorised with a wide use of printed circuitry making for compact size, simple installation and high reliability (both in and out of doors). High sensitivity and 625 line resolution ensure excellent picture quality under normal lighting conditions. Closed circuit television provides the penetrating, all-seeing eye that scans, inspects, controls and directs-that is today accepted as invaluable in almost every aspect of industry, commerce, transport and education. A wide range of accessories are available which further increase the system's

ALMOST LIMITLESS APPLICATIONS

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	CAMERA	DISTRIBUTION		CONTROL	 ÓISTRIBUTION	*******	S MUNITUR S		BASIC
		UNIT		MONITOR	UNIT	**********	VIEWING (STSILM
2	**************************************				,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,		MONITOR	*******	SYSTEM

SYSTEM SPECIFICATION Scanning standards: 625 line, 50 fields, 2: | interlace. Horizontal resolution: 600 lines. Bandwidth 8 Mc/s over complete system. Linearity: $\pm 2\%$ positional error. Geometry: $\pm 2\%$ of rectangle averaged over picture. Auto Sensitivity: over the range 60: 1 in light value—normal picture obtained with illumination of only 2ft. candles (50% subject reflectance) at lens aperture of f/2. Spectral Response: Panchromatic. Ambient Temperature: Max. temperature for all units -30 C. to +55 C. Power requirements 90/130 v. and 200/240 v. A.C., 50-60 c/s. Consumption: 45 watts including camera. Camera Lenses: Standard 16 mm. cine lenses with "C" mounts are normally used. Accessories: See under Camera and Control Monitor.



CAMERA

Totally enclosed dustproof unit only $3\frac{3}{4} \times 4 \times 10\frac{1}{2}$ in., weighing 4 lb. Finished in two-tone blue/grey. Vidicon tube. Automatic sensitivity control enables the camera to maintain full picture quality over a brightness range of 60:1. 625 line scanning standard 2: l interlaced, frame synchronised to mains supply. 600 lines horizontal picture definition with a bandwidth of 8 Mc/s. All supplies are obtained from the control monitor (consumption 5 watts).

CAMERA ACCESSORIES

Lenses: Superb quality 25 mm. (1 in.) f/1.8, "C" mount lenses made especially for this system are available, also a limited Lenses: Superb quality 25 mm. (1 in.) f/1.8, "C" mount lenses made especially for this system are available, also a limited quantity of motorised zoom lenses. Remotely Controlled Weatherproof Pan and Tilt Heads: Pan 340° at 6° per sec., Tilt \pm 50° at 4° per sec. 230/250 v., 50 c/s operated. Remotely Controlled Pan and Tilt for Indoor Use Only: Details as above. Weatherproof Camera Housing: Windscreen Wiper, 75 w. heater, internal circulation fan, mounting bracket for camera housing (the latter items are extras for the Weatherproof Housing).

Vds. of cable. PRICES FOR LENSES AND ACCESSORIES ON APPLICATION.

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CONTROL MONITOR

14-in. screen, overall size 16×14×18 in. (excluding Remote Control Unit on which Monitor is shown), weight 30 lb. Panel controls provided : Mains on/off, Contrast, Brightness, Remote Focus. Preset controls (under side panels) include: Frequency lock, Monitor height, Frame linearity, Camera height, Camera width, Auto sensitivity, Camera linearity, Cable correction Video gain, Beam current, Y shift, Electrostatic focusing for camera and monitor. Additional input: Video –100 mV peak white positive into 50 ohms; Synch. - 2 v. peak/peak negative. Output: 100 mV peak white positive; 2 v. peak/peak negative Ambient temperature range -30° C. to $+55^{\circ}$ C.

ACCESSORIES

Remote Control Switching Unit (shown under Control Monitor): Controls auxiliary functions at the camera. i.e. pan/tilt zoom, windscreen wiper, etc. Size 18×14×3 in., weight 8 lb.

Distribution Unit: Used for selecting the required picture from those available on the control monitors and distributing it to the appropriate viewing monitor. Size $19\frac{1}{2} \times 13\frac{1}{2} \times 8\frac{3}{4}$ in., weight 30 lb.

Viewing Monitors: These are conventional domestic type receivers-19 in. and 23 in. models available. ~~~~

> Owing to the complexity and limited quantity of units available this equipment is available to CALLERS ONLY.

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28209 28300 28300 28300 28300 28300 28300 28300 28300 28300 28300 2830000 2830000000000	SILICON \bigstar PLANAR \bigstar N.P.N. \bigstar P.N.P. All these types available	All the above untested packs have an average of 75% or more good semiconductors. Free packs suspended with these orders. Orders must not be less than the minimum amounts quoted per pack. P/P 2/6 Per Pack (U.K.)
25.766 25163 251730 257700 257700	$\begin{array}{c cccccc} 2N299 & 28131 & 2N696 & 2N2906 \\ 28501 & 28512 & 2N697 & 2N743 \\ 9N9411 & 28109 & 2N1507 & 28731 \\ \end{array}$	THESE VERY POPULAR UNTESTED BRAND NEW TRANSISTOR PACKS ARE STILL AVAILABLE.
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Selection from our listsNo.PriceNo.Price15 50 Unmarked Trans. Untested10/-24 Solar Cells Inc. Book of Instructions10/-26 17 Red Spot AF Transistors10/-26 17 Red Spot AF Transistors10/-27 18 28 33/-002428 29 3110/-29 30 10 Red Spot AF Transistors10/-29 4 40NP 10 Light Sensitive Cell91 4 4091/-29 5 81. Trans. 400 Mc/sBrand New Trans.92 5 81. Trans. 400 Mc/sBrand New Trans.93 5 80 44 0NP NP 10185 44 0NP NP 10185 44 0NP NP 10186 40NP NP N1881 956 42 0NP NP 10181 956 42 0NP NP N1881 956 12 6 BSY 26/7 Trans. New Unctated10/-82 5 BC 107-8-910/-84 25 BC 107-8-910/-82 5 BC 107-8-910/-<	FOR BETTER VALUE	25 BF V50-51-52 NPN SULION TRANSISTORS 10/- 25 2N706-A NPN SULION TRANSISTORS 10/-
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Wireless World, June 1968


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SUPER POWER ALLOY MAGNET These fantastic ex WD magnets weighing only 4lbs, will lift well over 100 lbs. Fitted with swivelled handle and keeper. Size 4in. x 3≩in. x 1≩in. Packed in original makers' cases of two. Price 30/- per pair, plus 7/6 P. & P.



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and test prods. **371-9.U** Post paid. Three other models available from stock. Descriptive

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220/240v. A.C. COOLING UNIT 2,300 r.p.m. 6in. blade size. Smooth powerful motor. All metal construc-tion. Continuously rated. Individually tested. Offered at fraction of maker's price, £2/15/-. P. & P. 7/6.

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SANGAMO WESTON Dual range voltmeter. 0-5 and 0-100 v. D.C. FSD 1 mA. In carrying case with tests prods and leads. 32/6. P. & P. 3/6.

AUTO TRANSFORMERS. Step up, step down. 110-200-220-240 v. Fully shrouded. New. 300 watt type, £3 each. P. & P. 4/6. 500 watt type, £4/2/6 each. P. & P. 6/6. 1,000 watt type, £5/5/- each. P. & P. 7/6.

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CONTRACT OC	ARGAIN SALE	; NEW STO	CK AT UN	BEATABLE	PRICES;
OC44, OC45, OC OC71, OC72 equi	SID now only ivalent	1/0 each 1/- each	1! .!	£3.0.0 pc	er 100
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OCP71 UNMARKED, U	INTESTED TH	ANSISTORS	TO CLEA	R	2/- each! 7/6 for 50!
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BY100 type rect: £7.10.0 per 100;	£50.0.0 per 1,0	00.	PRICE! C	NLY 2/6 ea	ch; 24/- doz.;
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.02ŏμF 1μF	3 volt 5μ1 350 volt 6μ1		25 volt 3 volt	$^{64 \mu F}_{64 \mu F}$	2.5 volt 9 volt
$1.25\mu F$	$\begin{array}{c cccc} 16 \text{ volt} & 8\mu \text{H} \\ 3 \text{ volt} & 8\mu \text{H} \\ 0 \text{ volt} & 8\mu \text{H} \\ \end{array}$		3 volt 6 volt	$100 \mu F$ $100 \mu F$ $100 \mu F$	3 volt 6 volt
$2\mu F$ $2\mu F$ $2\mu F$	70 volt 8µ1 150 volt 10µ	r	275 volt 25 volt	150µF 200µF	12 volt 3 volt
$2\mu F$ $2.5\mu F$	350 volt 16μ 16 volt 20μ 95 volt 20μ	F F	150 volt 3 volt	200μF 250μF 250μF	4 volt 2.5 volt
2.5µF 3µF 3µF	25 volt 20µ 3 volt 20µ 25 volt 20µ	F F F	9 volt 15 volt	$250 \mu F$ $320 \mu F$ $250 \mu F$	2.5 volt 9 volt
$3.2\mu F$ $3.2\mu F$	6.4 volt 25μ 64 volt 25μ 4 volt 25μ	F F	6 volt 12 volt	350μF 400μF 400μF	10 volt 2.5 volt
$4\mu F$ $4\mu F$ $4\mu F$	12 volt 25μ 12 volt 30μ 25 volt 30μ	F F F	6 volt 10 volt	400μF 500μF 500μF	4 volt 6 volt
$4\mu F$ $5\mu F$	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	F F	3 volt 9 volt	640μF 750μF	2.5 volt 12 volt
All at 1/- each. Min	xed Packets of 20 (our selection) 10)/		
0.001µF	500 volt 0.02	μF	600 a.c.	0.25μF 0.5μF	350 volt
$0.001 \mu F$ $0.002 \mu F$ $0.005 \mu F$	500 volt 0.1	ιF	350 volt	$0.5 \mu F$.	500 volt
All at 15/- per 100.	3/- per dozen. M	ixed Bags of 100	(our selection	n) 10/	
MULLARD POL	YESTER CAPA	CITORS. AL	L HALF P	RICE	
$0.0022 \mu F$ 400 volts $0.0018 \mu F$ 400 volts $0.0015 \mu F$ 400 volts	s	4d. 4d.	$0.15 \mu F$ $0.22 \mu F$ $0.27 \mu F$	160 volts 160 volts 160 volts	·· ·. 7d. ·· ·. 7d.
$\begin{array}{ccc} 0.0010 \mu F & 400 \text{ volts} \\ 0.001 \mu F & 400 \text{ volts} \\ 0.01 \mu F & 400 \text{ volts} \end{array}$	5 5 5		$0.056 \mu F$ $1 \mu F$	125 volts 125 volts	
VERY SPECIAL V	ALUE! Silver Mic	a. Ceramic. Poly	styrene Cond	lensers.	
Well assorted. Mixe	ed types and value	s. 10/- per 100.			
					······
RESISTORS. Giv 6/6 per 100 or 55/- Also 3 to 3 watt close	ve-away offer; D	fixed types and d values, 7/6 pe	l values, ‡ t	o ½ watt, er 1.000.	·
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clos WIRE-WOUND	re-away offer; M per 1,000. se tolerance. Mixe RESISTORS	dixed types and d values, 7/6 pe	d values, ‡ t r 100; 55/- p	o ½ watt , per 1,000.	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clos WIRE-WOUND 1 watt, 3 watt, 6 wa	re-away offer; M per 1,000. se tolerance. Mixe RESISTORS att, 6d . each; 7 wa	Xixed types and d values, 7/6 per utt and 10 watt,	d values, ‡ t r 100; 55/- p 9d. each. 20	o ½ watt, her 1,000. assorted, 10/	
RESISTORS. Giv 6/6 per 100 or 55/- Also $\frac{1}{2}$ to $\frac{3}{2}$ watt clos WIRE-WOUND 1 watt, $\frac{3}{2}$ watt, $\frac{6}{2}$ wa CONNECTING W 10yd., $\frac{1}{-}$; 100yd.,	re-away offer; M per 1,000. se tolerance. Mixe RESISTORS ttt, 6d. each; 7 ws VIRE. THIN, H 7/6; 500yd., 25/-	Aixed types and d values, 7/6 per utt and 10 watt, P.V.C. INSULA (post 4/6); 1,00	d values, ‡ t r 100; 55/- p 9d. each. 20 ATED Joyd., 40/- (p	to ½ watt, per 1,000. assorted, 10/	
RESISTORS. Giv 6/6 per 100 cr 55/- Also i to 3 watt clou WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAI	Ye-away offer; J per 1,000. se tolerance. Mixe RESISTORS att, 6d. each; 7 wa VIRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt, P.V.C. INSULA (post 4/6); 1,00 BOXED. RO	d values, ‡ t r 100; 55/- p 9d. each. 20 MTED D0yd., 40/- (p CK-BOTTC	no ½ watt, ner 1,000. assorted, 10/ nost 6/-).	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clos WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAI DY87 EABC60	Ye-away offer; J per 1,000. se tolerance. Mixe RESISTORS ttt, 6d. each; 7 ws TIRE. THIN, I 7/6: 5009d., 25/- ND-NEW AND 6/9 EX F/F 7/. EX F/F	Aixed types and d values, 7/6 per utt and 10 watt, V.C. INSULA (post 4/6); 1,00 BOXED. RC 51 86	d values, ‡ t r 100; 55/- p 9d. each. 20 UTED 109yd., 40/- (p 0CK-BOTTCO 6/9 6/9	bo ½ watt, per 1,000. assorted, 10/ bost 6/-). DM PRICES: PCL85 PCL85 PEF200	8/5
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt close WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EAC820 ECC82 ECC82	Ye-away offer; J J per 1,000. se tolerance. Mixe se tolerance. Mixe Mixe RESISTORS tt, 64. each; 7 wa TIRE. THIN, I 7/6; 5009d25/- ND-NEW AND 6/9 EY; 7/- EY; 7/- EY;	Aixed types and d values, 7/6 per ttt and 10 watt, (post 4/6); 1,90 BOXED. RCC 51	d values, ‡ t r 100; 55/- p 9d. each. 20 MTED 00yd., 40/- (p CK-BOTTC 6/9 6/9 7/1 10/6	o ½ watt, her 1,000. assorted, 10/ post 6/-). DM PRICES: PCL85 PCL86 PFL200 PL81	··· ·· 8/5 ··· · 8/5 ··· · 11/8 ··· · 10/1
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clor WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W loyd., 1/-; 100yd., VALVES. BRAJ DY87 EABC60 ECC82 ECC82 ECC83 ECL86	re-away offer; J per 1,000. se tolerance. Mixe RESISTORS ktt, 64. each; 7 wa TIRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND 6/9 EY 7/- BY 6/9 EY 7/4 BY 7/4 PA 7/1 POC 8/5 PCC	Aixed types and d values, 7/6 per att and 10 watt, ext. and 10 watt, (post 4/6); 1,00 BOXED. RC 51 57 57 78 74 74 75 74 75 74 75 74 75 75 76 76 76 77 76 76 77 76 77 76 77	d values, ‡ t r 100; 55/- p 9d. each. 20 ATED 10yd., 40/- (p CK-BOTTC 6/9 6/9 7/1 10/6 10/6	o ½ watt, her 1,000. assorted, 10/ bost 6/-). M PRICES: PCL85 PCL85 PCL85 PL84 PL84 PL84 	····· 8/5 ···· 8/5 ···· 11/8 ···· 10/1 ···· 8/5 ···· 8/5
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clow WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC82 ECC86 EF86 EF86	re-away offer; J per 1.000. se tolerance. Mixe RESISTORS Rtt, 6d. each; 7 was TRE. THIN, I 7/8; 500yd., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt, ext. and 10 watt, <	l values, ‡ t r 100; 55/- p 9d. each. 20 ITED 100/d., 40/- (r CK-BOTTCC 	o ½ watt, her 1,000. assorted, 10/ most 6/-). M PRICES: PCL85 PFL86 PFL86 PFL86 PL586 PL590. PL590. PY32 PY32	····· 8/5 ····· 11/8 ···· 10/1 ···· 8/5 ···· 8/5 ···· 12/5 ···· 12/5 ···· 12/5 ···· 12/5
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt cloudy WIRE-WOUND 1 watt, 3 watt, 6 watt 000000000000000000000000000000000000	fe-away offer; J J per 1.000, se tolerance. Mixe se tolerance. Mixe Mixe RESISTORS stt, 64, each; 7 ws rIRE. THIN, I 7/6; 500yd., 25/- ND-NEW AND 6/9 EY,	Aixed types and d values, 7/6 per d.tt and 10 watt, v.C. INSULA (post 4/6); 1,00 BOXED. RC BCS0 S2 S24	l values, ‡ t r 100; 55/- p 9d. each. 20 UTED D0yd., 40/- (p 6/9 6/9 7/1 10/6 10/6 10/6 8/5 10/1 8/5 9/10 8/5	o ½ watt, er 1,000. assorted, 10/ oost 6/-). M PRICES: PCL85 PFL200 PL36 PL53 PL53 PL53 PL50 PY381 PY800	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clos Mine ± to 3 watt clos WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ VALVES. BRAJ CC82 ECC82 ECC83 ECL86 EF80 EF83 EF183 EF184 A turther 10% disco	Ye-away offer; J per 1,000. se tolerance. Mixe RESISTORS ktt, 64. each; 7 wg TIRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND 6/9 EY; 7/4 PA 7/4 PA 7/1 PC 8/5 PCC 7/1 PC 7/1 PC 7/1 PC 9/5 PCI out, will be given of PA	Aixed types and d values, 7/6 per dt and 10 watt, .v.c. INSULA .(post 4/6); 1,00 BOXED. RC 86	d values, ‡ t r 100; 55/- F 9d. each. 20 ATED 00yd., 40/- (p CK-BOTTC 6/9 6/9 7/4 10/6 8/5 8/5 ny one type.	o ½ watt, her 1,000. assorted, 10/ post 6/-). DM PRICES: PCL85 PCL85 PFL200 PL58 PL58 PL58 PL58 PY800	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt close WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC82 ECC83 ECC86 EF183 EF183 EF183 EF183 A further 10% disco RECORD PLAYI	re-away offer; J per 1,000. se tolerance. Mixe RESISTORS Rtt., 64. each; 7 was RES. THIN, I 7/6; 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt, extra and 10 watt,	d values, ‡ t r 100; 55/- p 9d. each. 20 TED 10yd., 40/- (r CK-BOTTC 	o ½ watt, her 1,000. assorted, 10/ bost 6/-). M PRICES: PCL85 PFL86 PFL86 PFL86 PFL85 PL84 PFL85 PL84 PFL85 PFL8	
RESISTORS. Giv 6/6 per 100 or 55/- Also $\frac{1}{2}$ to 3 watt cloudy WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAI DY87 ECC82 ECC82 ECC83 ECL86 ECL86 EFF85 EFF85 A further 10% disco All with needles.	re-away offer; J per 1.000, se tolerance. Mixe RESISTORS stt, 6d. each; 7 way TIRE. THIN, I 7/6: 500yd., 25/- ND-NEW AND . 6/9 . 7/4: 500yd., 25/- ND-NEW AND . 6/9 . 7/4: PX . 7/4 . 8/5 . 7/1 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 15/-; Acos GPS	Aixed types and d values, 7/6 per dtt and 10 watt, v.C. INSULA · (post 4/6); 1,00 BOXED. RC BCS0 57 70 70 71 724 739 736 737 738 738 739 734 · · · · · · · · · · · · · · · · · · ·	l values, ‡ t r 100; 55/- p 9d. each. 20 UTED 00yd., 40/- (r 6/9 6/9 7/1 10/6 10/6 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 10/2 10/2 10/2 10/2 10/2 10/2 10/2	o ½ watt, her 1,000. assorted, 10/ host 6/-). DM PRICES: PCL85 PFL200 PL56 PL53 PL53 PL50 PY81 PY81 PY81 PY800 /-; Acos GP94	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clor WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC60 ECC82 ECC82 ECC82 ECC86 EF183 EF183 A further 10% disco RECORD PLAYI Acos GP67/29 Monc All with needles.	re-away offer; J per 1,000, se tolerance. Mixe RESISTORS Rtt. 64. each; 7 was TRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt, extra and 10 watt, extra and 10 watt, (post 4/6); 1,00 BOXED, RC 51 57 58 57 58 57 58 50 51 52 53 53 54 53 54 55 56 57 58 59 50 51 52	d values, ‡ t r 100; 55/- p 9d. each. 20 ATED 10yd., 40/- (r CK-BOTTCC 6/9 6/9 6/9 7/1 10/6 10/6 10/6 9/10 9/10 9/10 8/5 9/10 8/5 9/10 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 8/5 9/10 9/	o ½ watt, her 1,000. assorted, 10/ bost 6/-). M PRICES: PCL85 PC	8/5 8/5 10/1 8/5 8/5 12/5 6/6 12/5 6/9 6/9 6/9 8/1 6/9 8/1
RESISTORS. Giv 6/6 per 100 or 55/- Also † to 3 watt cloud WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC82 ECC83 ECC83 ECC83 EF184 A further 10% disco RESORD PLAYJ Acos GP67/29 Monc All with needles. TRANSISTORISE VEROBOARD.	e-away offer; J per 1.000, se tolerance. Mixe RESISTORS RESISTORS TRE. THIN, I 7/8; 5009d., 25/- ND-NEW AND 6/9 EY 7/4 PG 7/4 PG 7/4 PG 7/1 PG 7/1 PG 9/5 PC point will be given of 25. EX ACARTRIDGH 0.15/-; Acos GPG ED SIGNAL INJ All sizes in stoce	Aixed types and d values, 7/6 per att and 10 watt, BOXED. RO BOXED. RO BOXED. RO <td>d values, ‡ t r 100; 55/- p 9d. each. 20 ITED 00yd., 40/- (r 6/9 6/9 6/9 7/1 10/6 8/5 10/1 8/5 8/5 my one type. 10/ SIGN</td> <td>o ½ watt, her 1,000. assorted, 10/ most 6/-). PM PRICES: PCL85 PFL86 PFL86 PFL83 PL500 PTS3 PCB2 PY81 PY800 /-; Acos GP94 (AL TRACE)</td> <td></td>	d values, ‡ t r 100; 55/- p 9d. each. 20 ITED 00yd., 40/- (r 6/9 6/9 6/9 7/1 10/6 8/5 10/1 8/5 8/5 my one type. 10/ SIGN	o ½ watt, her 1,000. assorted, 10/ most 6/-). PM PRICES: PCL85 PFL86 PFL86 PFL83 PL500 PTS3 PCB2 PY81 PY800 /-; Acos GP94 (AL TRACE)	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt cloudy WIRE-WOUND 1 watt, 3 watt, 6 watt CONNECTING Watt, 1 (1) 1 vatt, 2 watt, 6 watt CONNECTING Watt, 1 (1) 1 vatt, 2 watt, 6 watt, 1 (1) VALVES. BRAJ DY87 EABC80 ECC82 ECC83 ECL86 EF83 EF83 EF83 EF84 A turther 10% disco RECORD PLAYI Accos GP67/29 Monc All with needles. TRANSISTORISE VEROBOARD. 2µin. × 1in. 0.15 2µin. × 3µn. 0.15	e-away offer; J per 1.000, se tolerance. Mixe RESISTORS stt. 6d. each; 7 wa PIRE. THIN, I 7/6: 5009d., 25/- ND-NEW AND 6/9 EY 7/4 PG 7/4 PG 7/4 PG 7/4 PG 7/4 PG 7/1 PC 9/5 PC 9/5 PC 9/5 PC stats Acos GPS ED SIGNAL INJ All sizes in stoc: matrix	Aixed types and d values, 7/6 per dtt and 10 watt,	l values, ‡ t r 100; 55/- p 9d. each. 20 ITED 00yd., 40/- (r CK-BOTTC 6/9 6/9 7/1 10/6 9/10 8/5 9/10 8/5 mp one type. 10/ SIGN 17in.×8‡in ofn.×8in,	o ½ watt, her 1,000. assorted, 10/ nost 6/-). PM PRICES: PCL85 PFL200 PL36 PL53 PL53 PY381 PY381 PY381 PY380	
RESISTORS. Giv 6/6 per 100 cr 55/- Also i to 3 watt cloudy WIRE-WOUND 1 watt, 3 watt, 6 wa 00NNECTING W VALVES. BRAI 0ya. 1/-; 100yd., VALVES. BRAI DY87 ECL80 ECC82 ECC83 ECL80 EF85 EF85 EF183 EF184 A further 10% disco Ailwith needles. TRANSISTORISE VEROBOARD. 2im.x 2im. 0.15 2im.x 2im. 0.15 3im. x 32im. 0.15	re-away offer; J per 1.000, set tolerance. Mixe RESISTORS stt, 6d. each; 7 was TIRE. THIN, I 7/6: 5009d., 25/- ND-NEW AND . 6/9 . 7/4: 5009d., 25/- . 7/4: 5009d., 25/- . 7/4: PA . 7/4 . 7/4 . 7/4 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 9/5 . 15/-; Acos GPG	Aixed types and d values, 7/6 per att and 10 watt,	l values, ‡ t r 100; 55/- p 9d. each. 20 UTED 00yd., 40/- (r 6/9 6/9 6/9 7/1 10/6 10/6 10/7 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 10/1 8/5 10/1	o ½ watt, her 1,000. assorted, 10/ host 6/-). DM PRICES: PCL85 PFL200 PL36 PL53 PL53 PX81 PY81	
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clor Also ½ to 3 watt clor WIRE-WOUND 1 watt, 3 watt, 6 wa WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAD DY87 EABC60 ECC82 ECC83 ECL86 EF86 EF86 EF88 FF183 A further 10% disco RECORD PLAYI Alw with needles. TRANSISTORISE VEROBOARD. 2½m.x 3½m. 0.15 2½m.x 3½m. 0.15 2½m. x 3½m. 0.15 3½m. x 3½m. 0.15 3½m. x 5½m. 0.15	e-away offer; J per 1.000, se tolerance. Mixe RESISTORS tt, 64. each; 7 was TRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per d.t and 10 watt, .v.c. INSULA .(post 4/6); 1,00 BOXED. RC 86 87 86 87 88 89 89 89 89 88 <td< td=""><td>d values, ‡ t r 100; 55/- p 9d. each. 20 TED 0dyd., 40/- (r 0dyd., 40/- (r CK-BOTTC . 6/9 . 70 . 10/6 . 10/6 . 10/6 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 10/ SIGN 17in. × 83in . 53in. × 33in . 33in. × 33in</td><td>o ½ watt, her 1,000. assorted, 10/ bost 6/-). M PRICES: PCL85 PCL85 PL53 PL53 PL54 PY80 PY81. PY82. PY80. YY80. (-; Acos GP94 (AL TRACE) . 0.15 matt . 0.1 matr . 0.1 matr . 0.1 matr</td><td> 8/5 8/5 10/1 8/5 </td></td<>	d values, ‡ t r 100; 55/- p 9d. each. 20 TED 0dyd., 40/- (r 0dyd., 40/- (r CK-BOTTC . 6/9 . 70 . 10/6 . 10/6 . 10/6 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 9/10 . 10/ SIGN 17in. × 83in . 53in. × 33in . 33in. × 33in	o ½ watt, her 1,000. assorted, 10/ bost 6/-). M PRICES: PCL85 PCL85 PL53 PL53 PL54 PY80 PY81. PY82. PY80. YY80. (-; Acos GP94 (AL TRACE) . 0.15 matt . 0.1 matr . 0.1 matr . 0.1 matr	8/5 8/5 10/1 8/5
RESISTORS. Giv 6/6 per 100 or 55/- Also ½ to 3 watt clow WIRE-WOUND 1 watt, 3 watt, 6 wa WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC83 EF184 A further 10% disco RECORD PLAYI Aim with needles. TRANSISTORISE VEROBARD, 2im. x3in. 0.15 3im. x3in. 0.16 3im. x3in. 0.16 SPECIAL OFFER Cutter and 5 Board Cutter and 5 Board Cutter and 5 Roard	e-away offer; J per 1.000. se tolerance. Mixe RESISTORS RESISTORS REE. THIN, I 7/8: 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt, .v.c. INSULA (post 4/6); 1,00 BOXED. RO BOXED. RO S6	d values, ‡ t r 100; 55/- p 9d. each. 20 ITED 00yd., 40/- (r 00yd., 40/- (r (************************************	o ½ watt, her 1,000. assorted, 10/ assorted, 10/ most 6/-). m PRICES: PCL85 PFL200 PL36 PL36 PL36 PF33 PL500 PT38 PT43 PCL85 PF13 PCL85 PCL85 PCL85 PCL85 PT13 PCL85	8/5 8/5 10/1 8/5 12/5 6/6 6/9 6/9 6/9 6/9 14/8 14/8 3/11 x 3/11 Terminal Pins.
RESISTORS. Giv G/G per 100 or 55/- Also i to 3 watt cloudy watt, 3 watt, 6 wa CONNECTING W VALVES. BRAJ DYS7 FABCR0 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 ECCS3 EFF86 EFF85 A further 10% disco TRANSISTORISE TRANSISTORISE VEROBOARD. 24 m. × 24 m. 0.15 24 m. × 24 m. 0.15 34 m. × 34 m. 0.15	re-away offer; J per 1.000, set tolerance. Mixe RESISTORS stt, 6d. each; 7 wa TIRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND . 6/9 Y7/7 BY . 7/4; PA . 7/4 PX . 7/4 PS . 7/1 9/5 PCI . 10/ Scos GPS ED SIGNAL INJ All sizes in stoci matrix	Aixed types and d values, 7/6 per att and 10 watt,	l values, ‡ t r 100; 55/- p 9d. each. 20 UTED 00yd., 40/- (r 6/9 6/9 6/9 7/1 10/6 10/6 10/7 8/5 10/1 8/5 10/5 10/1 8/5 10/1 10/2	o ½ watt, her 1,000. assorted, 10/ assorted, 10/ M PRICES: PCL85 PCL8	8/5 8/5 10/1 8/5 8/5 12/5 6/6 9/- 6/9 6/9 6/9 6/9 6/9 k KIT, 10/- rix 14/8 ix 3/11 x 3/11 Terminal Pins.
RESISTORS. Giv 6/6 per 100 or 55/- Also † to 3 watt clov WIRE-WOUND 1 watt, 3 watt, 6 wa CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC82 ECC83 ECC83 ECC83 EF184 A further 10% disco RESCORD PLAYI Acos GP67/29 Monc All with needles. TRANSISTORISE VEROBOARD. 2 jin. × 1 m. 0.1 f 2 jin. × 2 m. 0.1 f 2 jin. × 3 m. 0.1 f 3 jin. × 3 m.	e-away offer; J per 1.000, se tolerance. Mixe RESISTORS RESISTORS RIRE. THIN, I 7/6; 5009d., 25/- ND-NEW AND 6/9 EY 7/4 PG 7/4 PG 7/4 PG 7/4 PG 7/4 PG 7/1 PG 7/1 PG 9/5 PCI 9/5 PCI 9/5 PCI sum will be given of ER CARTRIDGH 0.15/-; Acos GPG ED SIGNAL INJ Matrix matrix	Aixed types and d values, 7/6 per dtt and 10 watt, BOXED. RO BOXED. RO BOXED. RO	1 values, ‡ ti 1 values, ‡ ti r 100; 55/- p 9d. each. 20 ITED 00yd., 40/- (p 00yd., 40/- (p CK-BOTTC 6/9 . 6/9 . 71 . 10/6 . 8/5 . 9/10 . 8/5 . 9/10 . 8/5 . 9/10 . 8/5 . 9/10 . 10/ SIGN 17in.× 3‡in. 3‡in.× 3‡in. 3‡in.× 3‡in. 3½in.× 3‡in. 3½in.	o ½ watt, her 1,000. assorted, 10/ assorted, 10/ post 6/-). press pr	8/5 8/5 11/8 10/1 8/5 12/5 12/5 12/5 12/5 6/9 6/9 6/9 6/9 6/9 14/8 IX 3/11 X 3/11 Terminal Pins.
RESISTORS. Giv 6/6 per 100 cr 55/- Also i to 3 watt cloudy watt, 3 watt, 6 wa 1 watt, 3 watt, 6 wa connectring watt, 6 wa connectring watt, 6 wa connectring watt, 6 wa valves. BRAI 0 yar PALVES. BRAI DY87 ECC82 ECC82 ECC83 ECC83 ECC83 ECC83 ECC83 EF86 EF85 FF183 FF184 A further 10% disco All with needles. TRANSISTORISE VEROBOARD. 2 in. × 3 in. 0.15 3 in. × 3 in.	θe-away offer; J J per 1.000, se tolerance. Mixe RESISTORS stt, 64, each; 7 wa rREE. THIN, I r/6: 5009d., 25/- ND-NEW AND . 6/9 . 6/9 EY . 7/6: 5009d., 25/- ND-NEW AND . 6/9 . 7/1 EY . 7/4 PX . 7/4 PS . 7/1 PC . 9/5 PC . 9/5	Aixed types and d values, 7/6 per att and 10 watt,	<pre>1 values, ‡ t r 100; 55/- p 9d. each. 20 UTED 00yd., 40/- (r 6/9 6/9 6/9 7/1 10/6 8/5 10/1 8/5 8/5</pre>	o ½ watt, her 1,000. assorted, 10/ host 6/-). DM PRICES: PCL85 PFL300 PL36 PL53 PL53 PV31 PY31 PY31 PY32 PY32 PY32 PY300 /-; Acos GP94 (AL TRACE) . 0.15 matr . 0.1 matr	8/5 8/5 10/1 8/5 12/5 6/6 12/5 6/9 6/9 6/9 6/9 14/8 IX 3/11 Terminal Pins.
RESISTORS. Giv 6/6 per 100 or 55/- Also † to 3 watt cloud WIRE-WOUND 1 watt, 3 watt, 6 watt CONNECTING W 10yd., 1/-; 100yd., VALVES. BRAJ DY87 EABC80 ECC82 ECC82 ECC82 ECC82 ECC83 ECC83 ECC83 EF184 A turther 10% disco RECORD PLAYI Acos GP67/29 Monc All with needles. TRANSISTORISE VEROBOARD. Zinx > 3in. 0.15 Zinx > 3in. 0.15 Zinx > 3in. 0.15 Zinx > 3in. 0.15 SPECIAL OFFEE Few only Multimete Orders by post to G. F. MILWA	e-away offer; J per 1.000, se tolerance. Mixe RESISTORS tit, 6d. each; 7 ws TRE. THIN, I 7/6: 5009d., 25/- ND-NEW AND	Aixed types and d values, 7/6 per att and 10 watt,	d values, ‡ t r 100; 55/- p 9d. each. 20 ITED 100; 40/- (p CK-BOTTC 6/9 6/9 7/1 10/6 8/5 10/1 8/5 10/1 8/5 10/1 8/5 10/1 10/- SIGN 17in. × 3jin. 3jin. × 3jin. 3jin. × 3jin. 3jin. × 3jin. 3jin. × 3jin. 2 per volt, 84 DRAYT[o ½ watt, her 1,000. assorted, 10/ assorted, 10/ pot 6/-). prosection of the second second prosection of the second second prosection of the second second second prosection of the second	

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Power supply, PP893/ GRC 32A; Filter D.C. Power Supply F-170/GRC 32A: Cabinet Electrical CY 1286/GRC 32A; Antenna Box Base and Cables CY 728/GRC; Mast Erection Kits, 1186/GRC; Directional Antenna CRD.6; Comparator Unit, CM.23; Directional Control CRD.6, 567/CRD and 568/CRD; Azimuth Control Units, 260/CRD. Test Set URM.44, complete with Signal Generator TS.622/U. MOTORISED ACTUATOR: 115 v. A.C. 400 c/s. single phase, reversible, thrust approx. 3 inches complete with limit switches, etc. Price $\pounds 2/10/$ - each, thrust approx. 3 inches com postage 5/- (ex equipment). Actuator Type SR-43: 28 v. D.C. 2,000 r.p.m., output 26 watts, 5 inch screw thrust, reversible, torque approx. 25 lbs., rating intermittent, price &3 each, post 5/-. FRACTIONAL MOTORS & FANS: Low inertia Motor 5UD/5361, Type 903, 24 v. input D.C., \$2/10/- each, 5/- post. Model PM84: 28 v. D.C. @ 2 amps., 4,500 r.p.m., output 40 watts continuous duty complete with magnetic brake. Price £2 each, postage 4/-. 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$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	AD140 AD149 AD161 AD162 AF114 AF115 AF116 AF117 AF118 AF212 ASZ21 BC107 BC108 BC109 BC108 BC109 BC184 BC742 BC743	8/- 8/- 8/- 8/- 4/- 4/- 4/- 5/6 4/99 4/66 5/6 5/6	188.20 5/6 + 0.8.72 1/6 + 0.6207 5/6 ZENERS: 3-30 Volt 14 watt, 5/6. Unmarked, untested transistors 60 for 10/ VEROBOARD: 24in. × 39in., 3/6. 24in. × 5in., 4/3. 37in. × 37in., 4/3. 37in. × 37in., 3/6. 24in. × 5in., 5/ ZEIN: × 39in., 3/6. 24in. × 5in., 4/3. 37in. × 37in., 4/3. 37in. × 37in., 16/6. 24in. × 50in., 5/ 10/ ELECTROLYTICS: 15 volt 1. MFD-100 MED., 1/6 each. 16 16/6. 1 watt, 9d. 3 watt, 1/6. 5 watt, 2/ SPEAKERS: 3 OHM 3in., 8/6. 5in., 14/6. Sin., 25/6. 12in., 39/6. 7in. × 4in., 16/6. Post and Packing, 1/- per order ALL DEVICES ARE FULLY GUARANTEED ALL DEVICES ARE FULLY GUARANTEED							
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R ADIOGRAMS, transistors and amplifiers repaired: you send small ones, we handle big ones any-mic Greater London-Tel. Hemei Hempstead 3915. Gregg Radio. Ltd., Hemei Hempstead. [243 R EPAIRS-Our modern service department equipped with the latest test equipment including a wow and flutter meter and multiplex stereo signal generator is able to repair. Hi Fi and tape recording equipment to manufacturers' standard.-Telesonic Ltd., 92. Tot-tenham Court Rd., London, W.1. 01-636 8177. [21

FOR SALE AND WANTED **ADVERTISEMENT FORM** TURN TO PAGE No. 129

ELECTRONIC ENGINEERS

Service Engineers required for Offices, throughout the United Kingdom, of well-known Company manufacturing Electronic Desk Calculating Machines. Applicants should possess a sound knowledge of basic electronics with experience in electronics, Radar, Radio and TV or similar field. Position is permanent and pensionable. Comprehensive training, on full pay, will be given to successful applicants. Please send full details of experience to

the Service Manager, Sumlock Comptometer Ltd., 102/108 Clerkenwell Road, London, É.C.1.



RADIO AND ELECTRONIC ENGINEERS BOARD OF TRADE (CIVIL AVIATION)

Qualified engineers required as Assistant Signals Officers in the field of Civil Aviation for the provision and installation of advanced electronic equipment-including the latest type of radar, telecommunications, navigational aids, etc.

QUALIFICATIONS: Degree with 1st or 2nd class honours in Electrical Engineering or Physics, or have passed all examinations for M.I.E.E., A.M.I.E.R.E. or A.F.R.Ae.S. AGE: 23 and normally under 35 on 31st December 1968 (extension for Forces and Overseas Civil Service).

SALARY (Inner London): On the scale £1,160-£2,092 depending on age and qualifications. Pensionable appointments, Good prospects of promotion.

APPLICATION FORMS from Civil Service Commission, Savile Row, London, W.1. X2AA, quoting S/85/ASO.

Vacancies in the SCIENTIFIC CIVIL SERVICE include: COMMUNICATIONS-ELECTRONIC SECURITY DEPARTMENT, Eastcote, Middx.

PHYSICISTS and ELECTRONIC ENGINEERS (graded E.O./A.E.O.) to be members of teams engaged on telecommunications research and development. Duties would be allo-cated where possible to suit the interests and aptitudes of individuals and might include development of communications equipment, exploitation of new techniques such as M.O.S. integrated circuits, designs of digital circuits, conducting and interpreting complex measure-ments of both audio and radio frequency signals, and the use of computers to aid various research tasks. Relevant experience essential.

QUALIFICATIONS: Degree, H.N.C., or equivalent in appropriate subject. Under 22, minimum qualification is G.C.E. in five subjects, including two Scientific/Mathematical subjects at "A" level or equivalent level.

SALARY (Outer London): E.O. (minimum age 26) £1,440—£1,809; A.E.O. £659 (at 18)— £878 (at 22)—£1,092 (at 26 or over)—£1,318.

Prospects of permanent pensionable appointments.

APPLICATION FORMS from the Recruiting Officer, Communications-Electronic Security Department, 8 Palmer Street, London, S.W.I.

PROOF AND EXPERIMENTAL ESTABLISHMENT Eskmeals, Bootle Station, Cumberland.

ELECTRONICS ENGINEER (graded E.O./A.E.O.) as Instrumentation Officer to assist in the supervision of the Instrumentation Group at Eakmeals, which is responsible for operational and data-processing techniques associated with a missile tracking radar and on-line digital computer system; velocity measurement by doppler radar and optical screen methods; various recording tasks using kinetheodolite, camera, UV and magnetic tape techniques; and timing of a variety of high speed events, e.g. fuse functioning and rocket firings. A knowledge of electronics and previous experience of instrumentation of field trials desirable. Minimum age of A.E.O. is 22.

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APPLICATIONS to Ministry of Defence, CE2(f) (AD), Northumberland House, Northumberland Avenue, London, W.C.2.

Government of Zambia REQUIRES TELECOMMUNICATIONS TECHNICIANS

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for the General Post Office, on contract for one tour of 36 months in the first instance. Commencing salary according to experience in scale Kwacha 2292 rising to Kwacha 3216 a year (\pounds Stg.1337 \pounds Stg.1876) plus Inducement Allowance of \pounds Stg.506 rising to \pounds Stg.615 a year. Gratuity 25% of total salary drawn. A supplement of \pounds Stg.175/215 a year is also payable direct to an officer's home bank account. Both gratuity and supplement are normally TAX FREE. Free passages. Quarters at low rental. Children's education allowances. Liberal leave on full salary or terminal payment in lieu. Contributory pension scheme available in certain circumstances. Special terms of service apply to serving civil servants including employees of the General Post Office whose applications *must* be sub-

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Candidates should have had a sound technical education and possess City and Guilds Certificates or equivalent in telecommunications. They should preferably have ten or more years training and practical experience in one or more of the following branches Carrier Systems; HF and VHF Radio; Telegraph Machines; Mixed duties covering these branches.

Apply to CROWN AGENTS, M. Dept., 4 Millbank, London, S.W.1., for application form and further particulars, stating name, age, brief details of qualifications and experience, and quoting reference M2K/62916/ WF

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We require qualified and experienced Technical Authors to work on the documentation of a variety of new projects covering a wide range of advanced Avionic Systems including Automatic Flight and Navigation Control systems, digital controlled electronic displays, and associated Test Equipment.

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Please write or telephone for application form to:-

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st. ± 226 , plus 30/- carr. .rans.suitable for running above on 230 v. mains price: $\pm 4/10/$ - (only available with unit).

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IND. UNIT 120//FLK-9 This is a digital display unit built into a 10in. rack panel 31in. high by 17in, deep. There are three low voltage number tubes plus 1 with special markings. These are mounted side by side on the front panel, giving number 0 to 9, §in. high and work off 6 v. Each number tube has 10 lamps and the size of each tube is 4m. long \times 11in. high \times 1m. wide. Each lamp has 2 transistors to control it 2N404 and 2N1184. These are mounted on plug-in boards, one board to each tube with 20 transistors on a board, i.e. each unit has 4 plug-in boards with a total of 80 transistors. These were made in 63 by Sylvania for USAF. In very good condition. Frice: 26, plus 10/- carr. Single tubes with transistor board and plugs, price 50/- each, plus 4/6 postage.

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British Solomon Islands REQUIRE TELECOMMUNICATIONS ENGINEER

for the Posts and Telecommunications Department, on contract for one tour of two years in the first instance. Commencing basic salary in scale equivalent to \pounds Stg. 1042 rising to \pounds Stg. 1993 a year, liable to British Solomon Islands Income Tax. In addition an allowance ranging from \pounds Stg. 716 to \pounds Stg. 1160, normally TAX FREE, will be paid direct by the British Government to an officer's bank account in the United Kingdom. Gratuity 25% of total salary drawn. Free passages. Terminal payment in lieu of leave. Generous Education Allowances. Contributory pension scheme available in certain circumstances. Candidates should possess the Final City and

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Guilds Certificate in Telecommunications and have a good knowledge of the operation and expansion of Strowger telephone exchange equipment. Membership of the Institute of Technician Engineers and broad experience of the installation and maintenance of H F Communications equipment, associated receivers and test equipment in a supervisory capacity will be an advantage.

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NCR requires additional ELECTRONIC, ELECTRO-MECHANICAL ENGINEERS and TECHNICIANS to maintain medium to large scale digital computing systems in London and provincial towns.

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Wymondnam, NK. [241] FERROGRAPH record/playback amplifiers, £15; power units for above, £5; extra carr. 12/6.— Bancroft, 71, Leysholme Cres. Leeds, 12. [237 SCHOMANDL frequency meter FD1, with FDM1 converter, 1KHZ to 900MHZ, in good order; £100 or best offer.—Box WW236, Wireless World.

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Applications are invited for the post of a Telecommunications Technician in the Borough Architect's Department for servicing ground to air equipment at Luton Airport. Applicants experienced in the servicing of Decca 424 Radar, Marconi AD 210C Direction Finder, Mufax facsimile reproduction equipment and I.L.S. equipment and holders of appropriate H.N.C. certificate preferred. The duties will involve shift working.

Commencing salary within Technicians Grades 4/5/6 (£1020—£1665 per annum) according to qualifications and experience. Housing accommodation considered. Reasonable removal expenses paid.

Forms of application may be obtained from the Chief Executive Officer and Town Clerk, Town Hall, Luton, Beds, to whom applications should be submitted as soon as possible.

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Previous experience as an instructor is not considered to be absolutely essential, but might well be an advantage. We are most anxious to find someone who has the ability and a real desire to teach fellow technicians.

Anyone interested in this vacancy is invited to send full details of his qualifications and relevant experience to Mr. D. D. Davies, Sumlock Comptometer Ltd., The Island, Uxbridge, Middlesex.

NORTH THAMES GAS

TECHNICAL ASSISTANTS TO THE COMMUNICATIONS OFFICER

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cases. The salary will be within the range £1,295-£1,535 per annum.

Applications stating age, qualifications and experience, should be addressed to:

Appointments Officer, North Thames Gas, 30 Kensington Church Street, London, W.8 quoting reference WW/3820.

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TRANSISTORS, Siemens BC109Cs for sale, guaranteed to specifications, HFE 240-900, 100OFF 2/ea., 100FF 2/6 ea., 10FF 3/- ea.; immediate delivery, c.w.o. 6d., c.o.d. 1/6.--P. Pinnock, 114, Keslake Rd. London, N.W.6. Tel. 01-370-4961. [242 FOR sale, cathode ray tubes; we have 606 tested cathode ray tubes of G.E.C.. Mazda, 20th Century and Mullard manufacture; £160 or near offer for the bot, or if desired a detailed list can be obtained from the following: D. W. H. Jackson, 25, Victorian Grove. Stoke Newington, London, N.16. [2022 WELL-KNOWN units by leading makers! Oscillofrom £20 to £195; second-hand but in good condition: a.c. mains operated; 31/2in screens, twin beams, long persistence C.R.Ts on some models; few only-mot to be missed.-(Dept. PRM), Lind-Air Optronics, Ltd., 25 & 53, Tottenham Court Rd., W.1. 01-580 1116/7 & 4532 & 7679.

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Please apply in writing stating the vacancy which interests you, giving full details of experience and career to date, to :

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HARWELL

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The minimum qualification for the post is five G.C.E. passes including English language at "O" Level and two scientific or mathematical subjects at "A" Level, but preference will be given to those with a degree (or equivalent). In addition candiaates should have experience in general experimental techniques involving vacuum apparatus and electronic and optical equipment; experience of electron spin resonance apparatus is desirable but not essential.

Starting salary will be at a point on the scales $\pounds 625 - \pounds 1,335$ or $\pounds 1,465 - \pounds 1,860$ depending upon age and experience. The successful candidate may be eligible for Authority. Housing or Assisted House Purchase.

Application forms and further details may be obtained from:

Appointments Section 'A,' A.5240/45, A.E.R.E., Harwell, DIDCOT, Berks.

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STICK TO YOUR CUSTOMERS! Self-Adhesive Service Labels with your NAME and PHONE NUMBER On all your Sales, Servicing, Repairs, etc. Size $2'' \times 1''$. Two colour. 3 day delivery. 35/- per 1,000. Samples: PERMARK (W), Station Grove, Wembley, Middx.

ASSISTANT Engineer

for CLOSED CIRCUIT TELEVISION

Assistant engineer required to maintain and operate 1" helical scan machines in Recording Centre of rapidly expanding C.C.T.V. Company. Salary to commence at £1,350 p.a. and rising after six months satisfactory service to £1,500 plus pension scheme.

Apply to:

The Technical Director, Television Applications Limited, 9-11 Windmill Street, London, W.1. Tel: Mus 3521.

LECTURER IN WORKSHOP PRACTICE

GOVERNMENT OF KENYA

Qualifications: Ordinary diploma in Mechanical/Electrical Engineering or its equivalent and a minimum of three years' experience teaching use of hand tools, lathes, machinery soldering, etc.

Duties: To instruct at engineering and tradesmen's levels all forms of workshop practice.

Age Limit: Up to 50 years.

Terms of Appointment: On contract for one term of 24 months at salary (subject to local income tax) in the scale £Kenya 1,128-1,440 (£Sterling 1,316-1,680). In addition directly paid Inducement Allowance (nontaxable) £Sterling 800-909 will be paid by British Government direct to officer's bank account outside East Africa. Terminal Gratuity 25%. Free family passages, etc.

Candidates, who must be nationals of the United Kingdom or the Republic of Ireland, should apply quoting RC 237/95/06 and giving full names, age, qualifications and experience, to :--

MINISTRY OF OVERSEAS DEVELOPMENT,

Room 301, Eland House, Stag Place, Victoria, London, S.W.1.

REDIFFUSION (EAST MIDLANDS) LIMITED, CASTLE BOULEVARD, NOTTINGHAM.

Senior Equipment Engineers are urgently required to be responsible for the maintenance of high quality sound and vision receiving site equipment; control equipment; high power audio amplifiers; and vision repeater equipment etc. associated with extensive H.F. Wired Vision Networks. Applicants should have a good knowledge of the principles and practice, of the transmission of vision and sound signals along cables; R.T.E.B. or City and Guilds examination qualifications are desirable but not essential. The salaries offered are from £1,100 and upwards depending upon the applicant's age, experience and qualifications and the Company may assist towards removal expenses etc. Please write to the Chief Engineer for application forms.

ULSTER: THE NEW UNIVERSITY SCHOOL OF HUMANITIES

LANGUAGE LABORATORY TECHNICIAN

To service, maintain and operate three 12 booth laboratories and associated equipment.

Applicants should have knowledge of solid state circuitry, radio/tape servicing and ability to use frequency test and measuring equipment.

Salary: £722-£1,007 or £987-£1,225

Applications with details of training and experience, and the names and addresses of two referees should reach the Registrar, New University of Ulster, Coleraine, Northern Ireland, by 31st May, 1968.

REDIFFUSION

TELEVISION FAULTFINDERS

We have vacancies for experienced television faultfinders in our Production Test Departments. R.T.E.B. Final Certificate or equivalent qualifications or experience are required, a knowledge of transistor circuitry will be an advantage. These positions will be staff appointments with all the expected benefits.

Applications to :

Works Manager, Rediffusion Vision Service Ltd., Fullers Way South, Chessington, Surrey (near Ace of Spades). Phone: 01-397-5411

B^{BC.2} television, radio, service, spares. Modify your set BBC.2. New manufacturer's conversion fits and tuners; list available. Pye integrated 405/ 625 transistorised tuner, £5/5; dual 405/625 IF transistorised panel incl. circuits, £2/15, p/p 4/6; Philips 625 conversion kit incl. tuner, 625 IF panel, switch assy. 7 valves, circuits, etc., £5/5 (or less valves 50/-), p/p 6/-; GEC/SOBELL 405/625 IF and output chassis incl. circuit, 86/6, p/p 4/6; FERGUSON 625 IF amp chassis incl. 6 valves, 55/- (or less valves 7/6), p/p 4/6. UHF tuners incl. valves, 38/6 (less valves 15/-), p/p 4/6, new VHF tuners, GEC transistorised, 70/-; A.B., Philips dual stand, Brayhead 3005, 30/-; Cyldon C, 20/-; KB 16 MC/s, 38 MC/s, 10/-, p/p 4/6. UMF tuners, used, good condition, 30/-, p/ 4/6; TV signal boosters, transistorised Pve/Laberat Bi/B3 or UHF battery operated, 75/-; UHF mains operated, 97/6; UHF mainstormers, Salvaged components-large selection of transformers, Salvaged components-large selection of transformers, Salvaged components-large selection of transformers, scant colls, turrets, etc. Enquiries invited, C.O.D. despatch available, --Manor Supplies, 64, Golders Manor Drive. Loncon, N.W.11. Callers 5899, High Ro. North Finchley, N.12 (near Granville Rd). Hursday 1.918 (day), set. 4032 (evg.). Early closing Thursday 1.918 (day). VIDEO and audio tape ¼in and lin×2,400ft, huge guantity available at fraction of original cost, please write for details; also 10/an Video presiston spools for ¼in or lin tape, 17/6 and 20/- respectively; Mullard Perrox cores LA1, LA2SO9, 7/6; LA5, LA6, LA7, 12/6; Plessey vibrators, type 1214, 10/-; Plessey loudspeakers 6in×4in, also 7in×4in, 35 ohms, 15/-; Plessey ganged potentiometers 20K+20K Linear, 7/6; valves N78 10/-, EN84 5/-; Hivac mains neons (bultin resistor) with 6in length twin lead, ideal for electronic gadgets and novelies, new reduced, price 1/3 each or 12/6 dozen (also available 110V); transistors. Mullard OC205 5/-, R.C.A. 2N410 2/6, CV2389 (OC71) 2/-; miniature germanium diodes, new and tested, 100 for 10/-; electrolytic capacitors, 125 MFD, 500V 4in×134, 9/-, 1000 MFD, 50V 2in×1354n 7/6, 50+50+ 50 MFD, 350V 3in×1354n 8/-. Enquiries invited for all electrolytics, very wide selection, discount on quantifited with sprung nylon locator, 1 pole-12 way, 3 pole-4 way, 4 pole-5 way, 2 pole-6 way, etc., 3/6 each, or 36/- dozen. Trv us for all "hard to get" spare parts; our prices are reasonable, our service is good and we are world-wide exporters and overseas agents, etc. Please let us know your immediate requirements; lists available.—Elekon Enterprises, 34, Baker St., London, W.1. 01-486 5553. [245]

AIR FORCE DEPARTMENT

- * INTERESTED IN DOING VITAL WORK ON R.A.F. RADAR AND WIRELESS EQUIPMENT?
- * Aged 19 and over and of good educational standard (G.C.E. "O" level passes in English language, Maths and Physics. City and Guilds Telecommunications Technicians Certificate or equivalent qualifications are desirable though not essential).
- * Experienced in radio/radar servicing.

IF SO, WE OFFER:

- * A first class opening as a Civilian Radio Technician. Present salary starting at up to £1,076 p.a. (according to age) and rising to £1,242, and good prospects of promotion (top posts in excess of £2,000 p.a.).
- * Five-day week. Three weeks 3 days annual leave at the start rising to 6 weeks, plus public holidays.
- * Excellent prospects of a good pension. If you do not qualify for a pension, then you receive a gratuity if you leave after at least 5 years service. Vacancies exist at R.A.F. Sealand, near Chester, but also arise from time to time at other R.A.F. Stations throughout the country.

MICROWAVE SYSTEMS Test Engineers

Pye Telecommunications Ltd. require at their factory at Haverhill, Suffolk, an Engineer to take charge of an expanding systems engineering team. There are also vacancies for Senior Engineers to become members of this team for work on production test of Broad Band Solid State Link equipment. Experience of video and/or multi-channel telephony is desirable, preferably with knowledge of semi-conductor work. Preference will be given to applicants holding a good academic qualification.

Attractive salaries will be offered and some assistance with housing in this expanding town may be possible.

All applications will be treated in the strictest confidence.



Apply in writing giving details to: The Works Manager

PYE TELECOMMUNICATIONS LTD.

Colne Valley Road, Haverhill, Suffolk,

Radiomobile

BRITAIN'S CAR RADIO SPECIALISTS

have a vacancy for a fully experienced

SERVICE ENGINEER

The successful applicant will be employed in our Main Service Workshop repairing:—

Transistorised & Valve operated Car Radios, Car Tape Recorders, Coach Radio & P.A. Equipment,

and also in our Service Garage on installation work and the servicing of equipment already fitted to vehicles.

After gaining considerable knowledge of our products, duties, in the future, may be extended to include mobile Field Service work.

This position carries a good commencing salary together with above average fringe benefits.

Applications should be made in writing to:

The Personnel Manager, RADIOMOBILE LIMITED.

Goodwood Works, North Circular Road, London, N.W.2. GLA 0171

A subsidiary of SMITHS INDUSTRIES LTD.

SCIENCE RESEARCH COUNCIL RADIO & SPACE RESEARCH STATION DITTON PARK, SLOUGH, BUCKS.

Experimental Officers/Assistant Experimental Officers are required to assist in work on the propagation of radio waves through the lower atmosphere. Duties will include the design and development of apparatus, its use in carrying out experiments and the analysis of the results obtained. Successful applicants may be required to work either on experiments using an 82 foot steerable aerial for studying VHF and UHF propagation or on the propagation of millimetre and sub-millimetre waves.

Qualifications:

University or C.N.A.A. degree, H.N.C. or equivalent qualification. If under age 22, five G.C.E. passes including two science or mathematical subjects at "A" level (or equivalent).

Salaries

Bucks.

A.E.O. between £584 and £1,243. E.O. between £1,365 and £1,734. Send for details to:— The Secretary, S.R.C. Radio and Space Research Station, Ditton Park, Slough,

CAMBRIDGE WORKS LIMITED Haig Road

JUNIOR ELECTRONIC ENGI-NEER required to join a small team developing test instruments for telecommunications. Previous experience of circuit design desirable, together with some mechanical skill. Staff appointment. $37\frac{1}{2}$ hour week. Age 21-25.

Please apply to the **Personnel Manager** in writing or by telephone, **Cambridge 51351, Ext 327.**

MINISTRY OF TECHNOLOGY CAREER OPPORTUNITIES FOR GRADE I CRAFTSMEN

Opportunities exist for Craftsmen in the Inspection Division on work involving inspection of electronic equipment. PAY for Grade I Craftsmen is £20.18.0 (London), £20.5.0 (Provincial) for a 40-hour, 5-day week. PROSPECTS: Promotion to higher grade posts with salaries up to £2352 for men who possess or obtain the necessary qualifications e.g. O.N.C. QUALIFICATIONS: A recognised apprenticeship or equivalent experience and training. HOLIDAYS: 2 weeks (80 hours) rising to 3 weeks (120 hours) after 5 years' service in the grade, plus 8½ days' public and privilege holidays. Applications, giving brief details of apprenticeship and/or experience, should be sent to: Industrial Personnel Officer, Ministry of Technology, Room 217, Baynards House, I-13 Chepstow Place, London W.2.

Inner London Education Authority

Norwood Technical College

Telecommunication & Electronics Department

Required for September for duties in the Department of Telecommunication and Electronics:

2 SENIOR LECTURERS 2 LECTURERS GRADE II 1 LECTURER GRADE I

1 LECTURER GRADE I Applicants, who would be concerned mainly with the work of the full-time Diploma and Certificate Courses, should have good general academic qualifications in Radio and Electronic Engineering and allied subjects. Recent high-level teaching particularly for the Senior Lecturer and Lecturer Grade II positions. Additionally, specialist knowledge of, and familiarity with modern techniques in, for instance, Computer Engineering, Microwave and U.H.F, techniques, Radio Transmission and Reception, Television Broadcasting, Radar and Radio Navigational Aids, would be an advantage. Participation in research and development projects is encouraged.

Assistance towards removal expenses may be given. Salary Scales:

Senior Lecturer; £2,280-£2,595. Lecturer Grade II; £1,725-£2,280. Lecturer Grade II; £1,035-£1,735, with additions for qualifications; starting point depends on qualifications, training and experience.

The London Allowance of £70 should be added to all the above scales.

Application forms, returnable within 14 days of the appearance of this advertisement, are obtainable from the College Secretary.

ENGINEERS Have you considered a career in Technical Authorship? If you have sound experience in electronics or communications and ability to write clear concise English we would train applicants as Technical Authors. The commencing salaries range from £1,300 to £1,700 depending on experience with the prospects of high future rewards and earnings.

Box No. 5039, c/o Wireless World

SENIOR STATION ENGINEER Education Department Government of Hong Kong

Duties; To advise on the detailed planning of the engineering requirements of the Hong Kong Government Television Studio complex, and to Supervise the day to day organisation of the Service and co-ordinate the transmitting of the television programmes. In addition to liase and advise on the installation of equipment and to train technical operations Staff and a local successor to take over the duties of the post.

Qualifications; Preferably under 45 years with either (a) B.Sc. (electrical engineering) or equivalent degree or (b) A.M.I.E.E. or (c) H.N.C. At least 10 years post qualification experience covering technical design, development and organisation operation and maintenance of a television broadcast complex and the operation and maintenance of videotape recording equipment and studio cameras (preferably Pilumbicon).

(preterably Financicol). **Terms of Appointment**: On contract for three years. Salary scale, \$4,5\$0 per month rising by annual increments of \$140 per month row \$5,000per month (\$3,77\$, \$3,894, \$4,009, \$4,125 p.a.). subject to local income tax. Terminal gratuity 17% free family passages, generous leave and education allowance.

Candidates should apply quoting RC 237/81/01 giving full names, age, qualifications and experience

The Appointments Officer, MINISTRY OF OVERSEAS DEVELOPMENT, Room 301, Eland House, Stag Place, Victoria, London, S.W.1.

ULTRA ELECTRONICS LTD.,

Urgently Require

TEST ENGINEERS

must be experienced in the testing and fault finding of complex electronic equipment.

PROTOTYPE WIREMEN

Applicants must be able to work from circuit diagrams and verbal instructions. Both vacancies offer a high rate of pay, good conditions, canteen social and sports club.

Write or phone:-

Personnel Officer, Ref. WWI, Ultra Electronics Ltd., Western Avenue, Acton, London W.3 Telephone: 01-992 3434.



are required to fill a number of vacancies at our Works in St. Albans and in Luton. The work is varied and will enable technicians with experience of r.f. circuits to broaden their knowledge of the latest techniques employed in the electronics and telecommunications industries by bringing them into contact with a wide range of the most advanced measuring instruments embracing all frequencies up to U.H.F.

The work involves calibration, testing and fault-finding on telecommunications measuring instruments.

Entrants may be graded as Testers, Test Technicians or Senior Testers according to experience and qualifications. Our expanding production programme geared to our recognised export achievement provides security of employment combined with good prospects of advancement not only within these grades but into other technical and supervisory posts within the Company.

Salaries are attractive and conditions excellent. The Company Pension Scheme includes substantial life assurance cover provided by the Company. Assistance with re-location may also be given in appropriate cases.



Please apply in writing to: **The Personnel Manager, Marconi Instruments Limited, c/o Directorate of Personnel (WW2890G), English Electric House, Strand, London, W.C.2.**

ELECTRONIC INSTRUMENT TECHNICIAN

We require a technician for the Instrument Section of our Engineering Group. Duties will include the design, construction and maintenance of specialised instruments for pilot scale food processing plant and our research laboratories. Location is Colworth House, near Bedford, which is a large food research centre supporting Unilever's food industries. Candidates should have HNC or equivalent. Starting salary will be determined by experience. The We Company has a Superannuation Scheme which includes life insurance. offer generous help towards relocation expenses for married men including removal expenses, legal and survey fees and help in obtaining house purchase loans.

Please write for an application form to the Assistant Staff Officer, UNILEVER RESEARCH LABORATORY, Colworth House, Sharnbrook, Bedford.



We have vacancies for Fault Finders, Testers, and Inspectors to work on in-teresting and advanced equipment includ-ing H.F. SINGLE SIDEBAND, V.H.F. RADIO TELEPHONES, U.H.F. MINI-ATURE EQUIPMENT.

Transistor experience is essential. Vacan-cies exist at all levels and training will be given where necessary.

Apply: Personnel Manager, CAMBRIDGE WORKS LTD., Haig Road, Cambridge.

CITY OF BIRMINGHAM EDUCATION COMMITTEE

MATTHEW BOULTON TECHNICAL COLLEGE DEPARTMENT OF ENGINEERING, Sherlock Street, Birmingham, 5.

FULL-TIME COURSE IN RADIO, TELEVISION AND ELECTRONICS

The proposed course is intended for young men who are employed in or are intent on making a career in the Radio/Television/ Electronic Industries, particularly in the field of servicing.

The course will be of 9 months' duration, starting in September 1968, and will lead to examinations held by the C.G.L.I. and R.T.E.B.

Further details and application forms may be obtained from the Head of Department. E. L. Russell, Chief Education Officer.

TECHNICIANS MINISTRY OF TECHNOLOGY **Requires Technicians**

Are you interested in electrical, electronic, or mechanical engineering? If so, there are excellent opportunities for you in the Ministry of Technology. The work involves the testing of radar, telecommunications apparatus, electrical power and navigation equipment, as well as the calibration of mechanical and electrical measuring devices.

These posts are mainly in the Woolwich, Harefield and Bromley areas, but vacancies also exist in other parts of the home counties and the U.K.

If you have an Ordinary National Certificate or a final City and Guilds Technicians Certificate you may well be the type of person we need.

The starting salary is £1,004 (age 24) rising by annual increments to £1,149 (age 28) and thence on to £1,283 with additional allowances for the London area and good prospects for promotion. There are also a few posts in the salary range £1,283 to £1,490 for well qualified and experienced candidates.

If you are interested, please send a post card to Mr. A. G. Stewart, Ministry of Technology, Aquila, Golf Road, Bromley, requesting an application form.

SCIENCE RESEARCH COUNCIL

RADIO & SPACE RESEARCH STATION DITTON PARK, SLOUGH, BUCKS

MALE ASSISTANT **EXPERIMENTAL OFFICERS**

are required for a three-year tour of duty at Stanlay, Falkland Islands, for operating and maintaining advanced apparatus for recording scientific information transmitted by telemetry from satellites (shift work). Married staff live in rent-free accommodation in modern well-furnished bungalows. Furnished bed-sitting rooms are available for single staff well-furnished bungalows. Furnished sitting rooms are available for single staff.

Oualifications:

University or C.N.A.A. degree, H.N.C. or equivalent qualification. If under age 22, five G.C.E. passes including two science or mathematical subjects at 'A' level (or equivalent) equivalent).

Salary

Salary whilst overseas on scale £690 per annum years £869; at age 26 years or over, £1,142. Additional overseas allowances and shift pay.

Apply:-

The Secretary, S.R.C. Radio and Space Research Station, Ditton Park, Slough, Bucks.

R ADIO-MIKES (Lustraphone). 2 units hardly used. Ist £155 each, accept £60 each, a genuine offer— Thornton's Electronics. Ltd., 18-19. Freeman St. Birmingham, 5. Tel. Midland 5588. [2029
 B'OXES of B.A. nuts and bolts, all brand new and high grade machine cut items, invaluable to all service men, experimenters, etc.; bolts include 2BA. 4BA and 6BA up to 2ln long, various heads, mainly brass, approx. 3-400 items per box; our special price f 6, plus 2/- post and packing.—Walton's Wireless Stores, 55a, Worcester St. Wolverhampton. [7]
 A Better deal for each customers. We do not provide

Stores, 55a, Worcester St., Wolverhampton. [71] A Better deal for cash customers. We do not provide interest free credit but offer a generous discount of 15% for cash. Equipment despatched brand new in sealed cartons on receipt of remittance with order. Agents for all leading makes. Demonstrations, service, guidance.-Write or 'phone. Callers welcome. Open all day Saturday, Thursday half day.-Audio Services, Ltd., 82. East Barnet Rd., New Barnet, Herts. Tel. Barnet 6605. [20]

ARTICLES WANTED

WANTED, televisions, tape recorders, radiograms, new valves, transistors, etc.—Stan Willetts, 37, High St., West Bromwich, Staffs. Tel. Wes. 0186. [72 WANTED, Eddystone communication receivers.-H.P. Radio Services, Ltd., 51, County Rd., Liver-pool, 4. Tel. Aintree 1445. [73 WANTED, all types of communications receivers and test equipment.—Details to R. T. & I. Electronics. Ltd., Ashville Old Hall, Ashville Rd., Lon-don, E.11. Ley, 4986.

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CENTRALLY situated in Portsmouth: modern ware-house/distribution centre of approximately 5,000 sq feet plus a workshop of 700 sq feet which is suitable for light industrial use; to bet on a long lease at £1.500 per annum exclusive: modern living accommodation is also available if required.— Please apply by jetter to the Secretary, 3, Victoria Crescent, Portsmouth. [2012



We have vacancies for . . .

Engineers experienced in the Maintenance and Operation of PROFESSIONAL TAPE RECORDING EQUIPMENT. A thorough knowledge of up-to-date audio techniques is essential.

Also vacancy for Instrument Maker/ Mechanic.

Salary according to qualifications, age and experience.

> **Apply: Chief Engineer REDITUNE LTD.,** Cray Avenue, Orpington, Kent. Tel: Orpington 32121

TEST EQUIPMENT ENGINEERS

Applications are invited from experienced engineers to fill these vacancies which arise in our Industrial Products Group located at NEWHOUSE, Scotland.

As a Test Equipment Engineer you will—

- ★ Develop, commission and install new test equipment.
- Check, calibrate and test simple and complex electrical and mechanical equipment.
- ★ Consult with and advise engineers and production personnel in use of equipment.
- ★ Assist in design of new test equipment and techniques.
- ★ Report and recommend on design and components.

Applicants should preferably be qualified to H.N.C. Standard with 2 years related experience or have had 5 years experience of special purpose production test equipment.

Assistance will be given with housing and removal costs where necessary.

Applications should be made, in writing, to:---

Personnel Manager HONEYWELL CONTROLS LIMITED Newhouse Industrial Estate MOTHERWELL, Lanarkshire, Scotland

Honeywell



Computer Test Technicians

Continuing expansion of our new factory at Winsford, Cheshire, enables us to offer interesting and varied work to test technicians at all levels. The work involves the testing of electronic devices and the servicing and calibration of test equipment.

If you have related experience in electronics, telecommunications or television, possibly acquired in H.M. Forces, we should like to hear from you.

Please write or telephone for application form: Assistant Personnel & Training Manager, English Electric Computers (Ref. WW.5181), Winsford, Cheshire. Tel: Winsford 3456.

BBC HV

YOUR HOBBY COULD BE YOUR PROFESSION

BBC Television employs many Film Recordists and Assistant Film Recordists in its Film Operations Department.

For young men between eighteen and twenty-eight, who are interested in imaginative, high-quality tape recording and are dedicated enthusiasts in the art of sound recording, there is a comprehensive training scheme which leads to a challenging and worthwhile career. Conditions of employment are generous and congenial and there are opportunities for travel at home and abroad.

For further details and application form, write to Head of Appointments Department, P.O. Box IAA, London, W.I. (quoting reference 68.G.662.W.W.).

UNIVERSITY OF SOUTHAMPTON

Department of Chemistry

Applications invited for the post of Technician in the Instrument Section to assist in the servicing of elec-tronic instruments and in the development of new equipment. While training will be given in the hand-ling of specialised equipment, previous electronics and electrical experience is essential. Qualifications to O.N.C. level or equivalent desirable but consideration will be given to those with a suitable background in practical electronics. Salary on scale £692 rising to £977. Pension scheme.

Applications should be sent to the:

Deputy Secretary, The University, Southampton, S09 5NH.

Giving the names of two referees preferably previous employers.



Contracts exist for the following authors: HAMPSHIRE: Electronic authors with HAMPSHIKE: Electronic authors with digital and analogue experience. ESSEX: Electronic authors with radar and solid state circuit experience, also Electro/Mechanical author with aircraft

experience. Please reply stating qualifications, experience and remuneration required to:

CALTHORPE ASSOCIATES, Manor Place, Tabors Hill, Gt. Beddow Chelmsford, Essex Tel.: Chelmsford 72872.

FOR sale, rebuilt television tubes, business plant or complete equipment.—Box WW2026, Wireless World.

NEW GRAM AND SOUND EQUIPMENT

GLASGOW.-Recorders bought, sold, exchanged; cameras, etc., exchanged for recorders or vice-versa.-Victor Morris, 343. Argyle St., Glasgow, C.2. [1]

VALVES

VALVE cartons by return at keen prices; send 1/-for all samples and list.-J. & A. Boxmakers. 75a, Godwin St., Bradford, 1. [10]

VALVES WANTED

W^E buy new valves, transistors and clean new com-ponents, large or small quantities, all details, quotation by return.--Walton's Wireless Stores, 55, Worcester St., Wolverhampton.

CAPACITY AVAILABLE

A IRTRONICS, Ltd., for coil winding, assembly and wiring of electronic equipment, transistorised sub-unit sheet metal work.—3a, Walerand Rd., London, S.E.13. Tel. 01-852 1706.

SHEET metal work, chassis, cabinets, engraving one off or production runs.—Apply Olson Elec-tronics, Ltd., Factory No. 2. 5-7, Long St., London, E.2. Tel. 01-739 2343. [2024]

TRANSFORMERS, chokes and all types of coil wind-ing, one off or production runs.—Apply Olson Electronics, Ltd., Factory No. & 5-7, Long St., Lon-don, E.2. Tel. 01-739 2343. [2025]

TUITION

KINGSTON-UPON-HULL Education Committee. College of Technology. Principal: E. Jones, M.Sc.,

F.R.I.C., FR.I.C., FULL-TIME courses for F.M.G. certificates and the Radar Maintenance certificate.—Information from College of Technology, Queen's Gardens, Kingston upon Hull.

RADIO officers see the world. Sea-going and shore appointments. Trainee vacancies in April and September. Grants available. Day and boarding students. Stamp for prospectus. Wireless College. Colwyn Bay. [12]

STUDY radio, television and electronics with the vorld's largest home study organisation, I.E.R.E., City & Guilds, R.T.E.B., etc. Also practical courses with equipment No books to buy. Write for free prospectus to ICS (Dept. 442), Intertext House, London, S.W.11. [24]

Tv and radio, A.M.I.E.R.E., City & Guilds, R.T.E.B., To certs., etc. on satisfaction or refund of fee terms; thousand of passes; for full details of cxams and home training courses (including practical equipment) in all branches of radio, TV. electronics, etc., write for 132-page handbook-free; please state subject.—British Institute of Engineering Technology (Dept. 150K), Aldermaston Court, Aldermaston, Berks. [15]

ENGINEERS.—A Technical Certificate or qualifica-tion will bring you security and much better pay. Elem. and adv. private postal courses for C.Eng., A.M.I.E.R.E., A.M.S.E. (Mech. & Elec.). City & Guilds, A.M.I.M.I., A.I.O.B., and G.C.E. Exams. Diploma courses in all branches of Engineering-Mech., Elec., Auto, Electronics, Radio, Computers, Draughts, Building, etc.—For full details write for FREE 132-page guide: British Institute of Engineer-ing Technology (Dept. 151K), Aldermaston Court. Aldermaston, Berks. [14

RECEIVERS 'AND AMPLIFIERS-SURPLUS AND SECONDHAND

HRO Rx5s, etc., AR88, CR100, BRT400, G209, S640, etc., etc., in stock.—R. T. & I. Electronics, Ltd., Ashville Old Hall, Ashville Rd., London, E.11. Ley, [65]

TEST EQUIPMENT — SURPLUS AND SECONDHAND

SIGNAL generators, oscilloscopes, output meters, wave voltmeters, frequency meters, multi-range meters, etc., etc., in stock.—R. T. & I. Electronics, Ltd., Ash-ville Old Hall, Asnville Rd., London, E.11. Ley. 4986. [64

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CITY & GUILDS (Electrical, etc.), on "Satisfaction or Refund of Fee" terms. Thousands of passes. For details of modern courses in all branches of elec-trical engineering, electronics, radio, T.V., automation, etc.; send for 132-page handbook-free.-B.LET. (Dept. 152K), Aldermaston Court, Aldermaston, Berks. [13]

BECOME "Technically Qualified" in your spare time, guaranteed diploma and exam. home-study courses in radio, TV, servicing and maintenance. R.T.E.B., City & Guilds, etc., highly informative 120-page Guide-free.—Chambers College (Dept. 837K), 146, Holborn, London, E.C.I. 116

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Tweeters 15 ohm, 65/-, P.P. 5/ As above less tweeters 3 or 15 ohm, 45/- ea., P.P. 5/	0.47µF, 2/8d. MODULAR METALLISED: P.C. mounting 250V:
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