

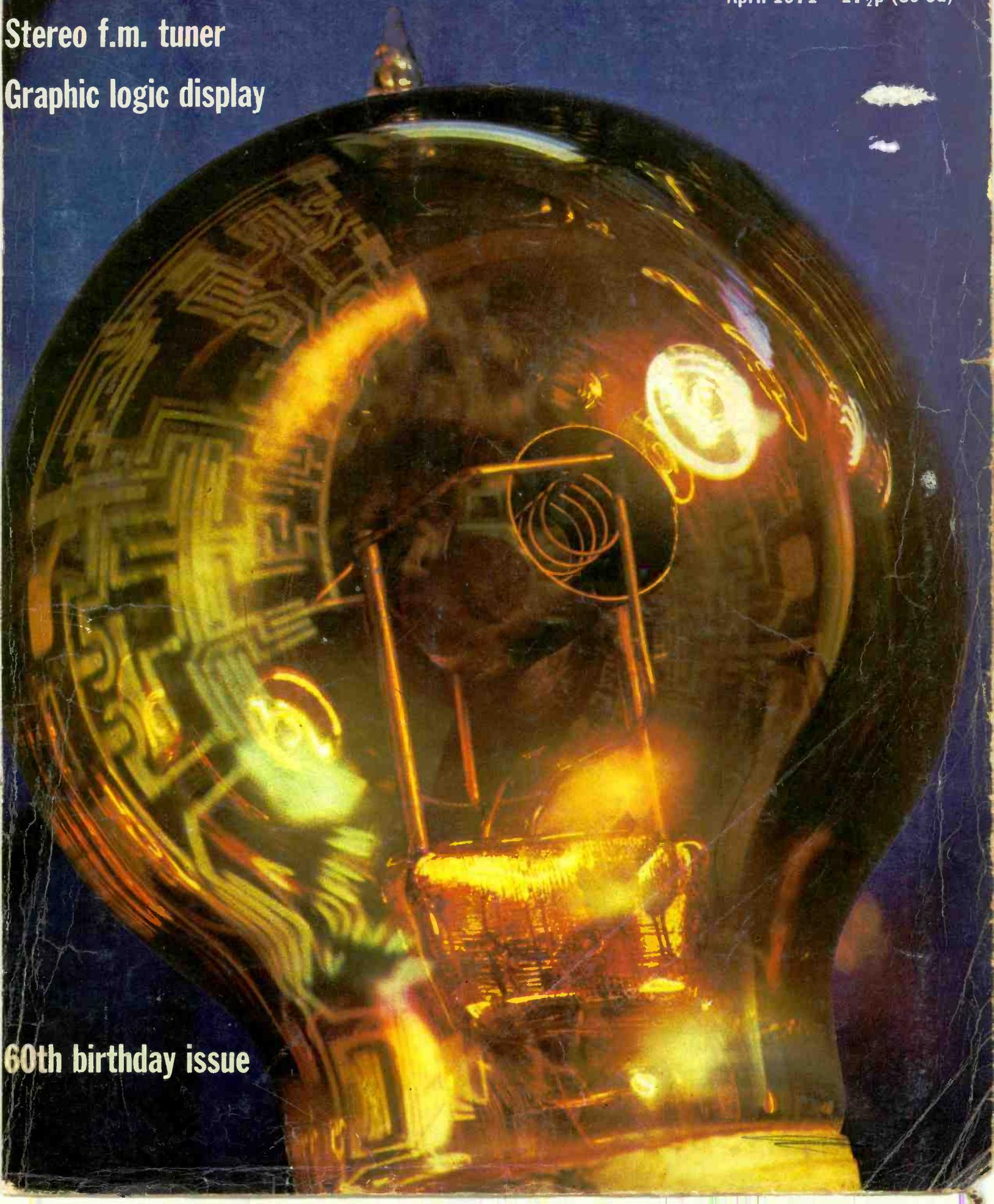
# Wireless World

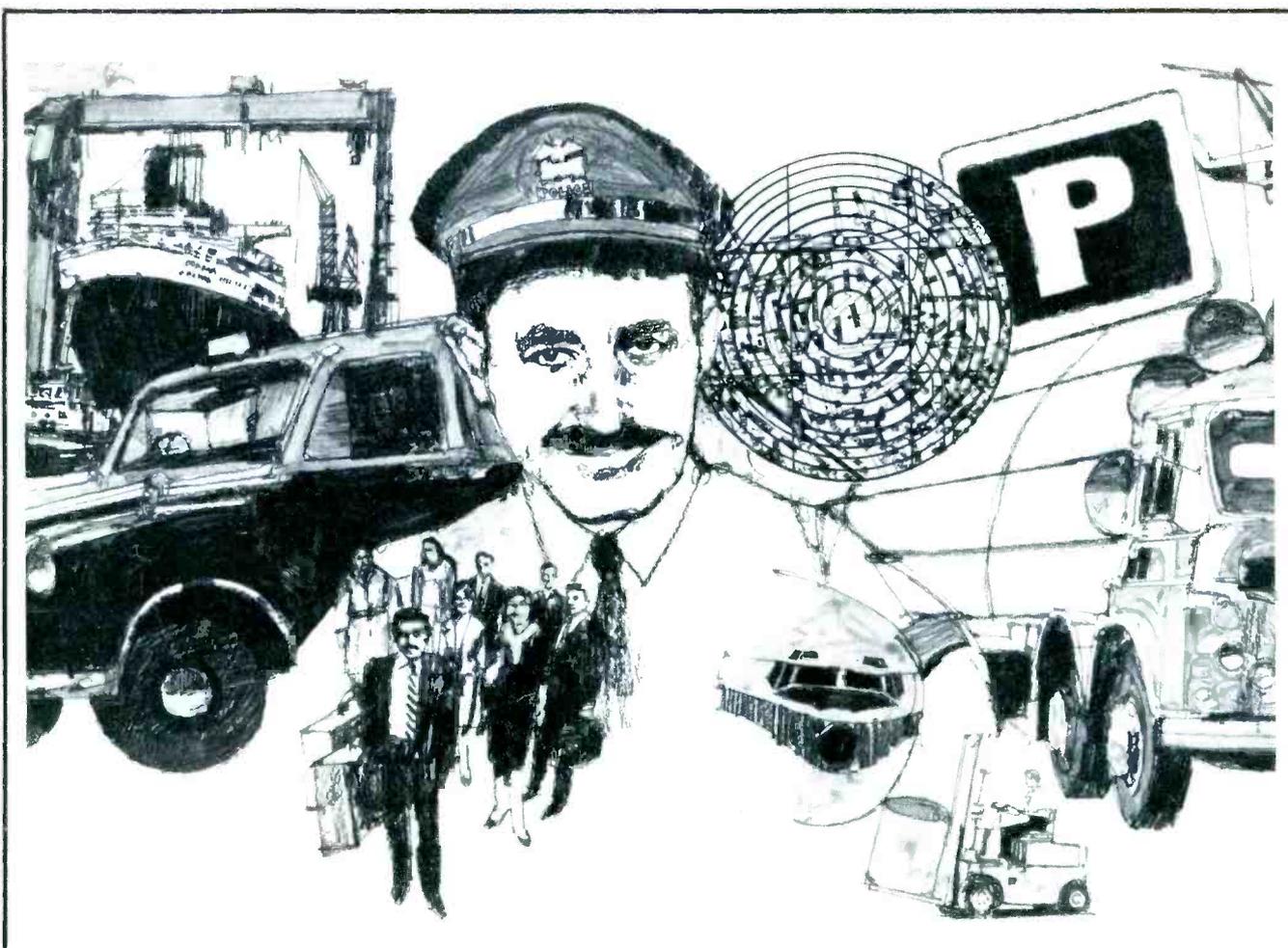
April 1971 17½p (3s 6d)

Stereo f.m. tuner

Graphic logic display

60th birthday issue





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**ITT** **Mobile**

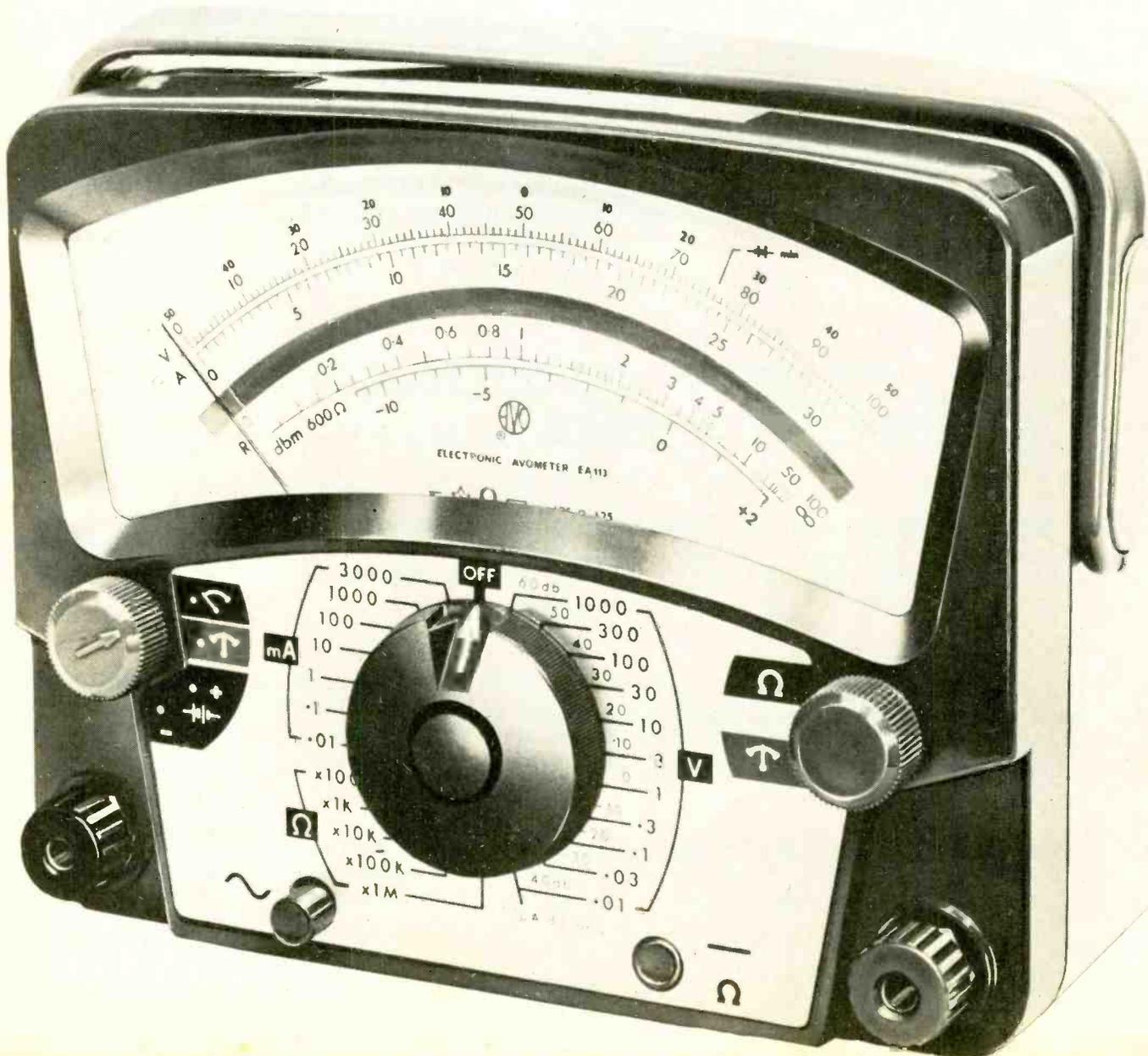
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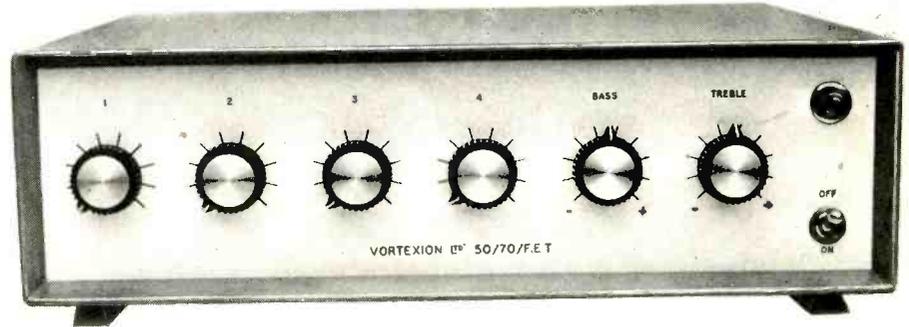
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**100 WATT ALL SILICON AMPLIFIER.** A high quality amplifier with 8 ohms—15 ohms or 100 volt line output for A.C. Mains. Protection is given for short and open circuit output over driving and over temperature. Input 0.4 V on 100K ohms.

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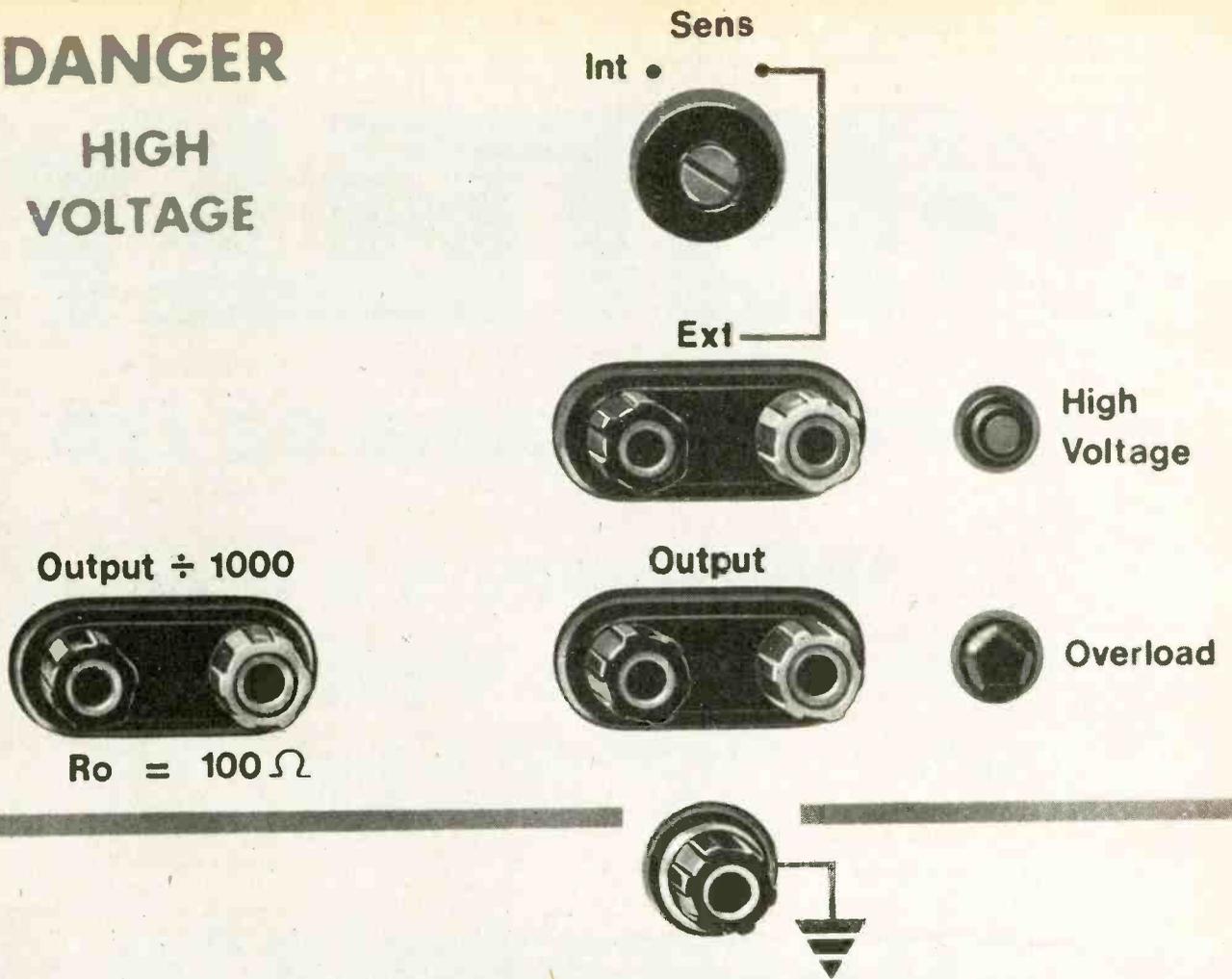
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## Confidere. v, (Lat.)

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(Founded 33 years ago)

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In electronics it is the noun RELIABILITY that is regularly used, and here the Dictionary refines the definition to "that may be relied upon; of sound and consistent character or quality".

However, to relate reliability only to the product is not enough. You, the Customer, have the right to expect reliability from every sector of your Suppliers' organisation, as well as from the products you buy. We therefore look on reliability as having to apply to our Company as a whole . . . reliability in research, design, manufacture, administration and in delivery as well as reliability in standards of performance of our products.

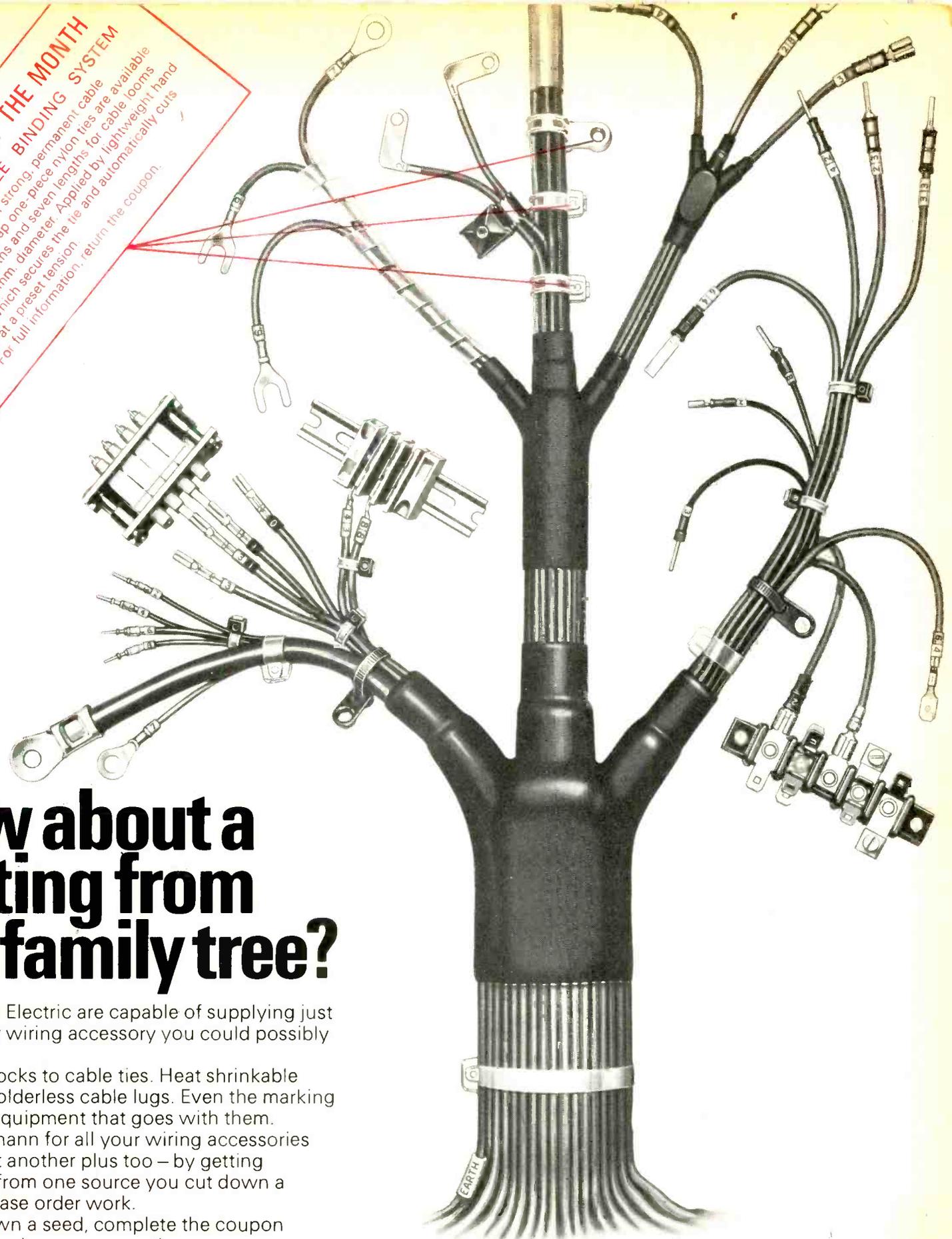


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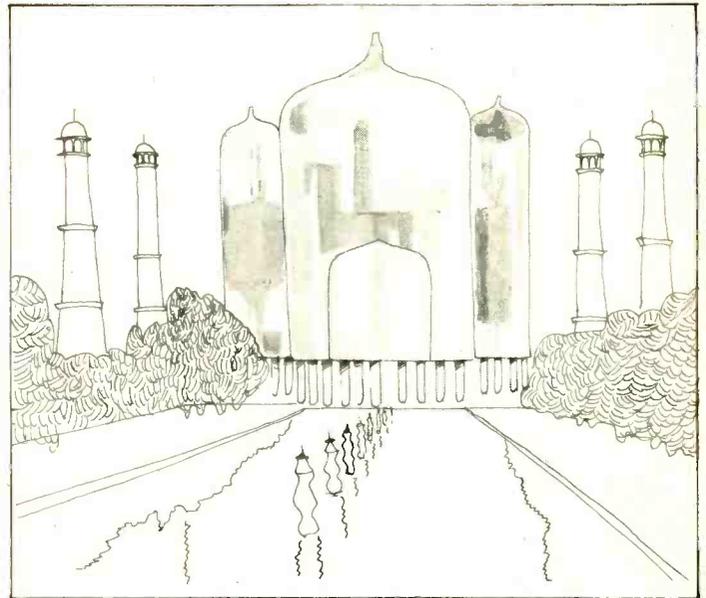
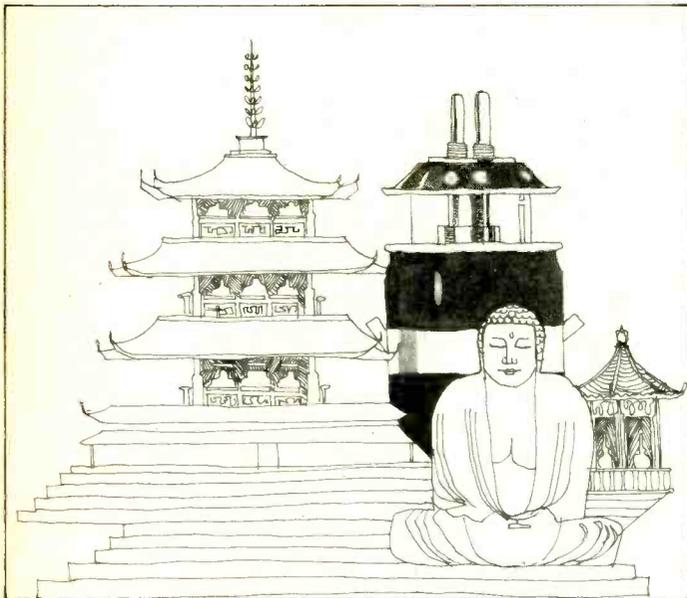
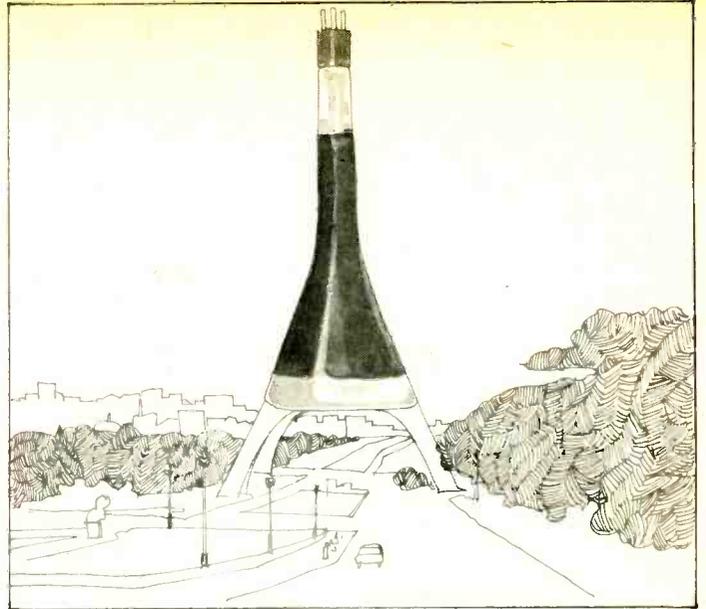
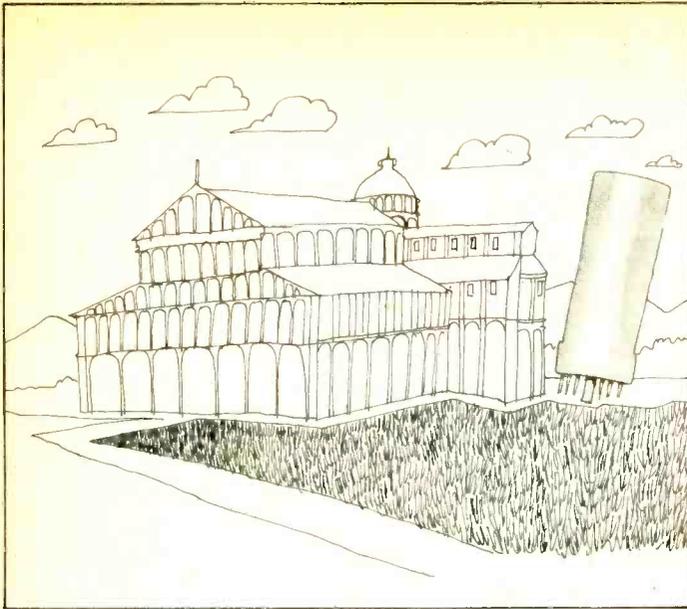


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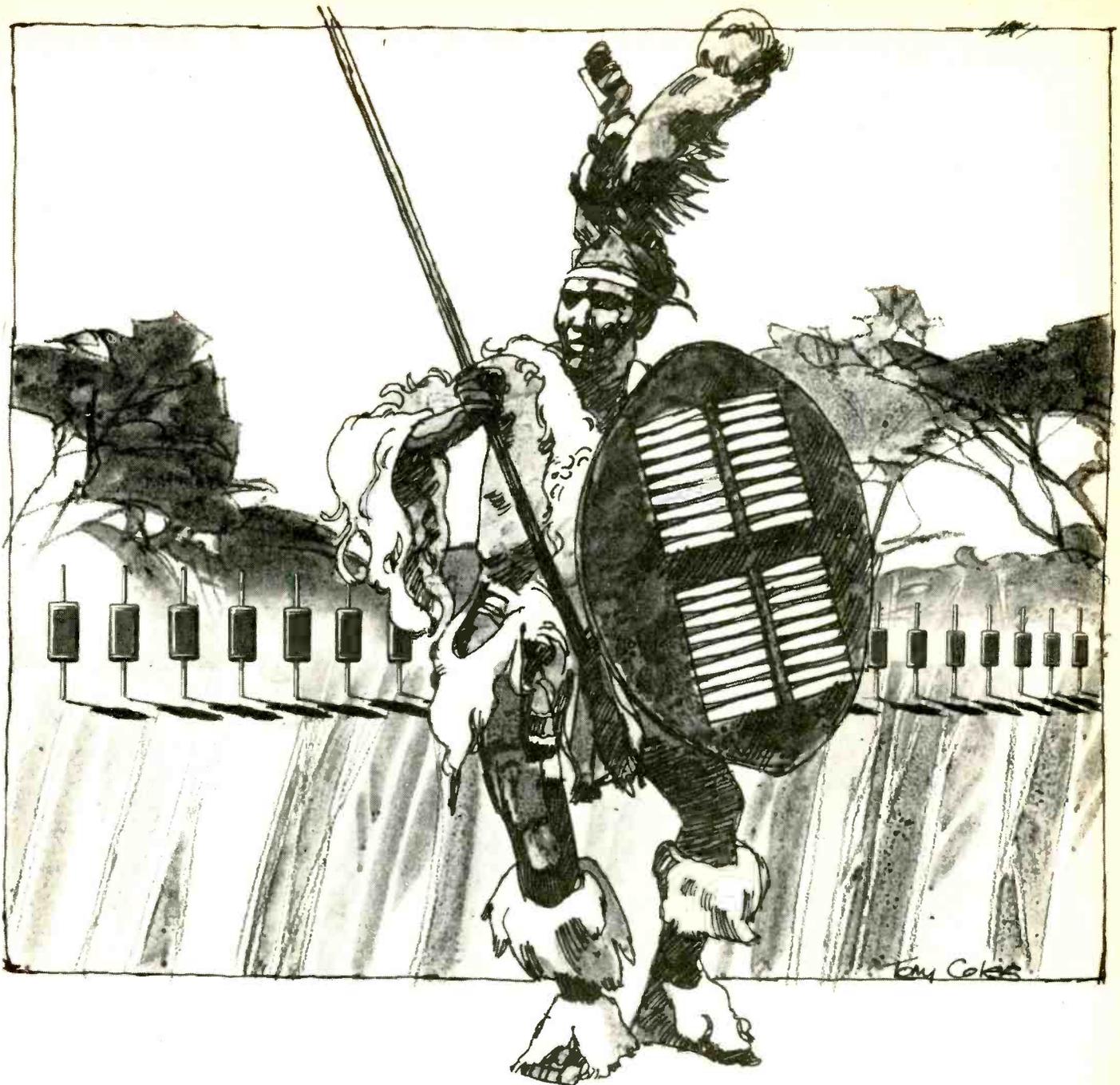
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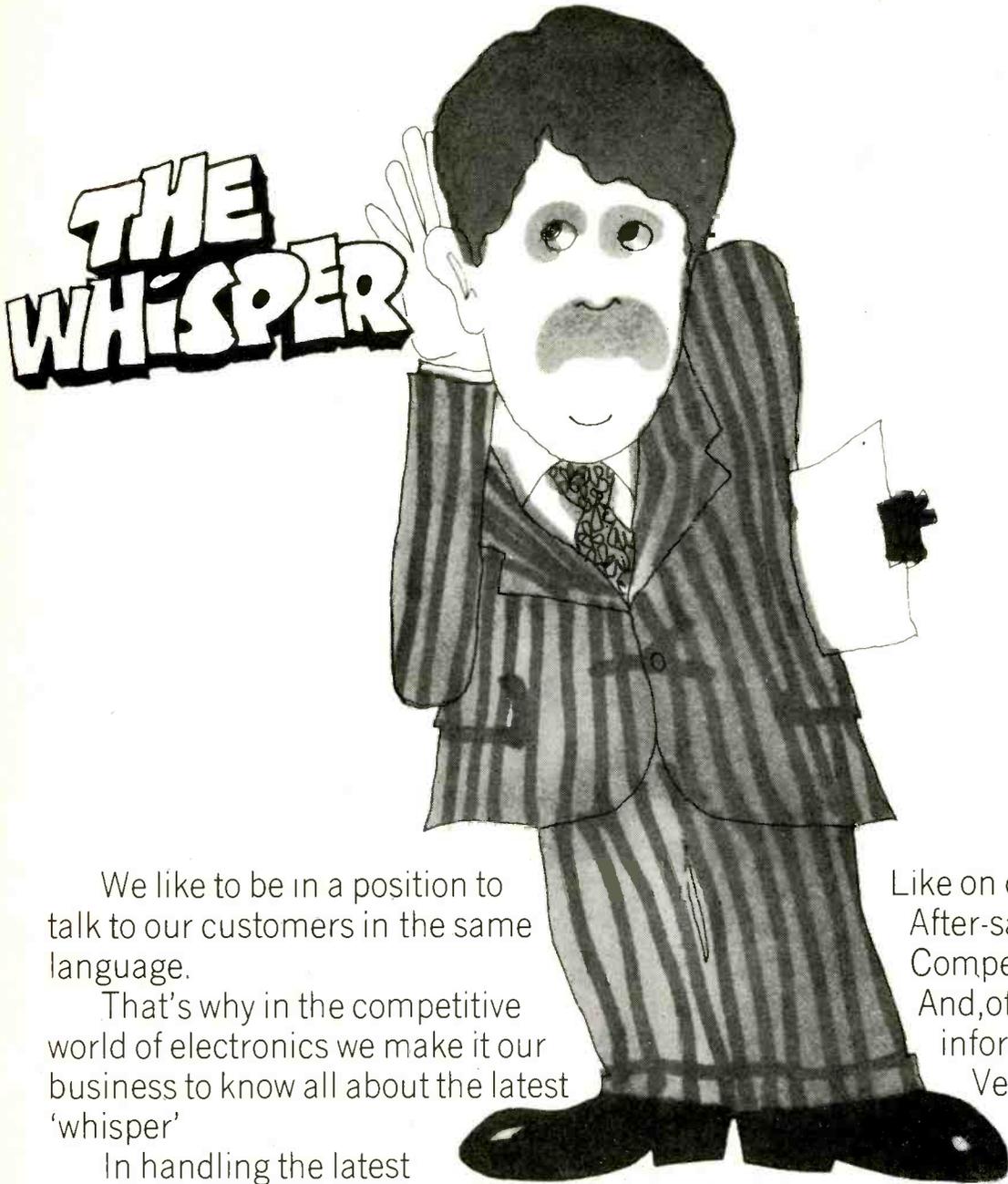
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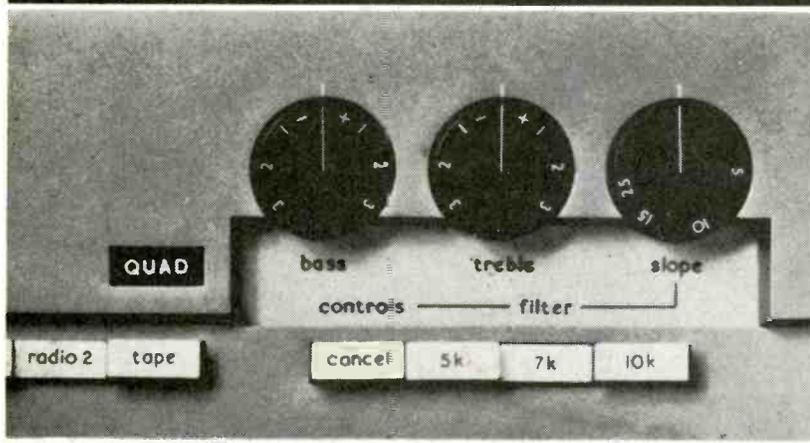
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**WEL**

## numbers



Set up a QUAD 33 with +1 on the treble control, and you will obtain a response precisely defined; readily and accurately repeatable. This response has a shape rather different from most run of the mill tone controls and there are, as you may guess, good reasons for this.

Then as the listener is not expected to know just what a given response curve does to the signal off the record, we provide a button marked 'cancel'. This enables him to make a direct comparison with the original and so learn just which recording defects need what correction. A QUAD user gets the best out of every record — every time — and enjoys the music to the full.

See and hear QUAD at SONEX '71—  
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# QUAD

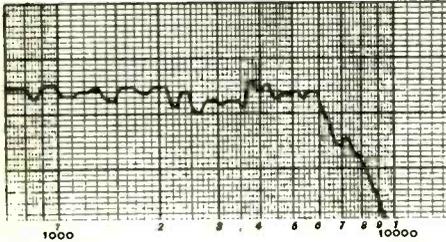
for the closest approach to  
the original sound

Send postcard for illustrated leaflet to Dept. WW  
Acoustical Manufacturing Co. Ltd., Huntingdon, Tel: (0480) 2561. QUAD is a Registered Trade Mark.

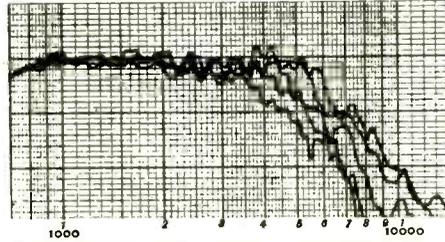


# Acoustic Research has measured the response of more than a million high-fidelity speakers.

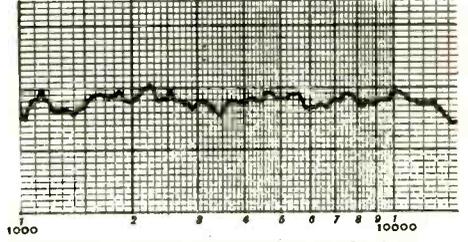
Here are some things we have learned about listening.



1. The frequency response of a midrange driver unit of an AR-3a, on axis. This corresponds to what one would hear outdoors, listening directly in front of a speaker.

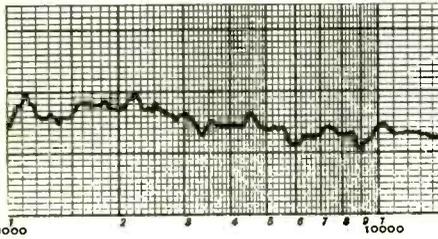


2. What happens when a listener moves over to one side of the speaker in 15° increments.

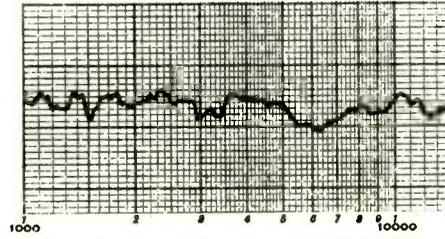


3. The integrated power output of the AR-3a above 1000 Hz, measured in a special reverberant chamber. Reflection from the walls of the chamber mixes together all of the sound emitted by the speaker system in all directions, an effect much more like that of a listening room than the anechoic chamber used for 1 and 2. A speaker system which measured well in both types of chamber would be accurate under almost all listening conditions.

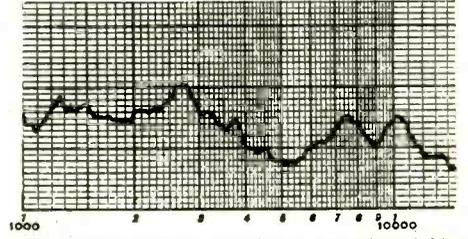
Integrated power output curves.



AR-3a and AR-5 with high-priced magnetic cartridge. It is interesting to see that the cartridge introduces somewhat more degradation of the signal than the speaker system, at least in the frequency range observed. Nevertheless, a small adjustment of the amplifier treble control could restore uniformity of response.



AR-2ax with moderately-priced magnetic cartridge. Although not as accurate as the AR-5 or AR-3a the AR-2ax displays the same kind of performance, that is, its integrated power output curve is relatively level. Because its dispersion, especially in the lower midrange, is less uniform the AR-2ax is more dependent on optimum placement than the others.



A 'multi-directional' system and a very expensive cartridge. Such systems are designed to take advantage of room reflections to smooth response and create spatial effects.

Vertical divisions  $\frac{1}{2}$  dB

## Fidelity means accuracy.

Accuracy distinguishes high-fidelity speaker systems from the speakers in simple radios and gramophones. It is therefore reasonable that evidence of accuracy should take precedence over descriptions of a speaker system's size, shape or theory of design. Acoustic Research offers exact measurement data for AR speaker systems to all who ask for it: music listeners, audio enthusiasts, science teachers, even competitors.

The accuracy of a speaker system can be evaluated by listening tests or by measurement. Both methods give the same information in different ways.

## Testing for accuracy.

To perform a listening test, an extremely accurate recording must be made and played back alongside the original source of sound. Amplifier and speaker system controls are adjusted to obtain as close a match as possible; and the speaker system judged by the degree of similarity. Acoustic Research has presented public concerts at which the Fine Arts Quartet and other musicians could be compared with recordings played back through AR speaker systems; even seasoned critics were deceived. Obviously, listening tests cannot be made with commercial recordings of music since the listener has no way of knowing which adjustment is most accurately reproducing the recording.

## Objective measurements.

While it is not always convenient to carry out scientifically controlled listening tests, properly conducted measurements can give the same information in permanent, quantitative form. AR knows something about this, having already tested the response of well over a million speakers — every one that we have ever made, and many made by competitors. Our findings are that the most important measurements required to assess the accuracy of a speaker system are (1) frequency response on-axis, (2) frequency response off-axis, (3) integrated power output.

AR speaker systems start at £39.95. Write to Bell & Howell for more information and a list of dealers.



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Colour TV  
Practical Radio & Electronics (with kit)

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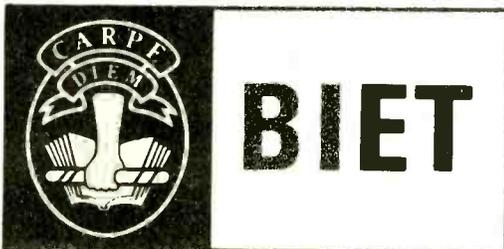
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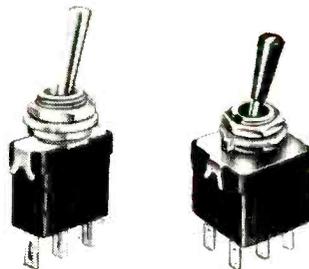


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subsidiary of. 

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# Two Schools of

Today, two schools of thought exist where there is a need for automated measurements.

# 1

## BUY THE SYSTEM

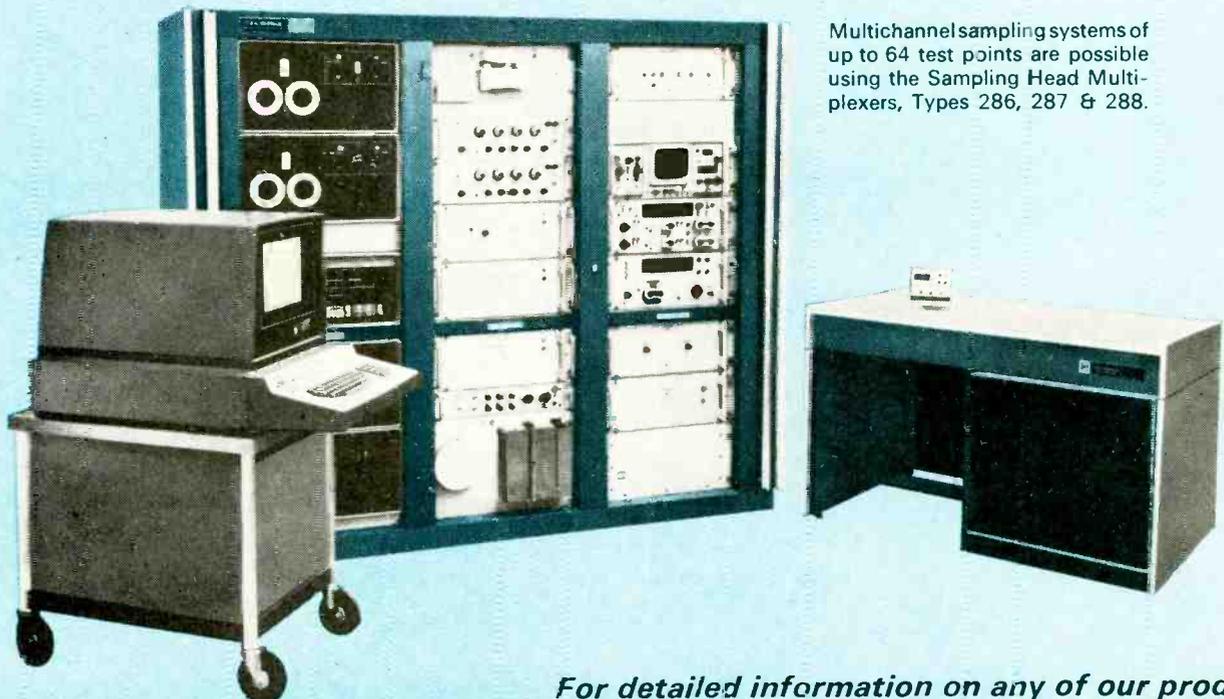
Buy a fully compatible test system and let the supplier do the systems engineering.

*Tektronix offers solutions . . . buy the system . . . for each approach:*

If you're of the "buy the system" group . . . you must decide: Functional, DC or Dynamic tests? The NEW Tektronix S-3150 will do *all three* types of testing on Digital IC's. Our S-3130 will make *dynamic* tests on IC's, transistors, circuit boards, etc. Most suppliers provide DC or Functional test systems . . . Tektronix provides dynamic systems!

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# Thought

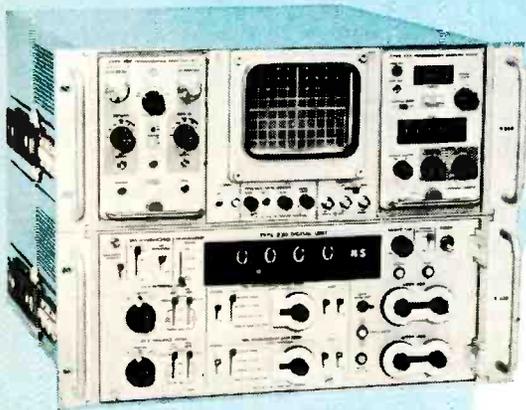
...from **TEKTRONIX**

# 2

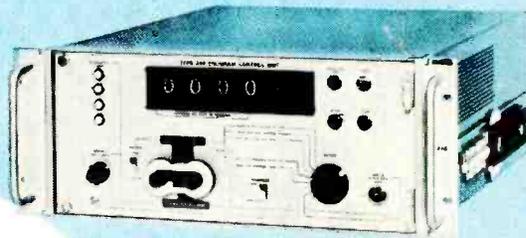
## BUY THE MODULES

Buy the best of each system module and do your own systems design.

*or buy the modules!*

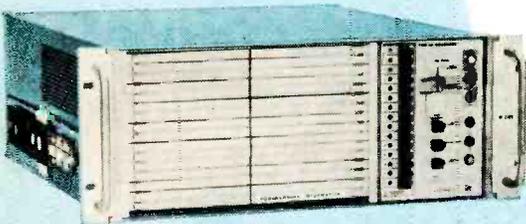


The Type 568/230 Digital Oscilloscope provides digital readout of measurements that are displayed in analog form on the CRT. With programmable plug-in units, all the measurement functions of the 568/230 can be externally programmed.

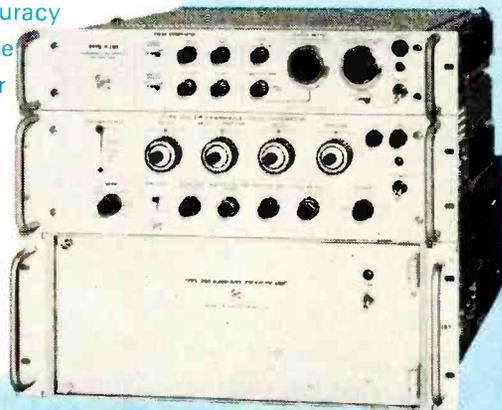


The Type 240 Programme Control Unit, in conjunction with a disc memory, can programme the 568/230 at speeds up to 100 measurements per second.

Prefer to buy the best systems modules available? Tektronix offers a digital oscilloscope with programmable plug-in units, multiplexers, programmers, programme control units and programmable pulse generators. Dynamic, switching-time measurements are made with greater speed, accuracy and convenience when you tailor your system around these modules.



The Type 241 will automatically sequence through 15 programmes, stopping on out-of-limits measurements. Programmes are easy to set up and change.



The Type 250 Auxiliary Programme Unit provides additional programming capabilities to the 240 and buffering for pulse generators, power supplies, etc.



I.L.E.C. SHOW STAND 1-115/131

# TEKTRONIX

committed to progress in waveform measurement

### Tektronix U.K. Ltd.

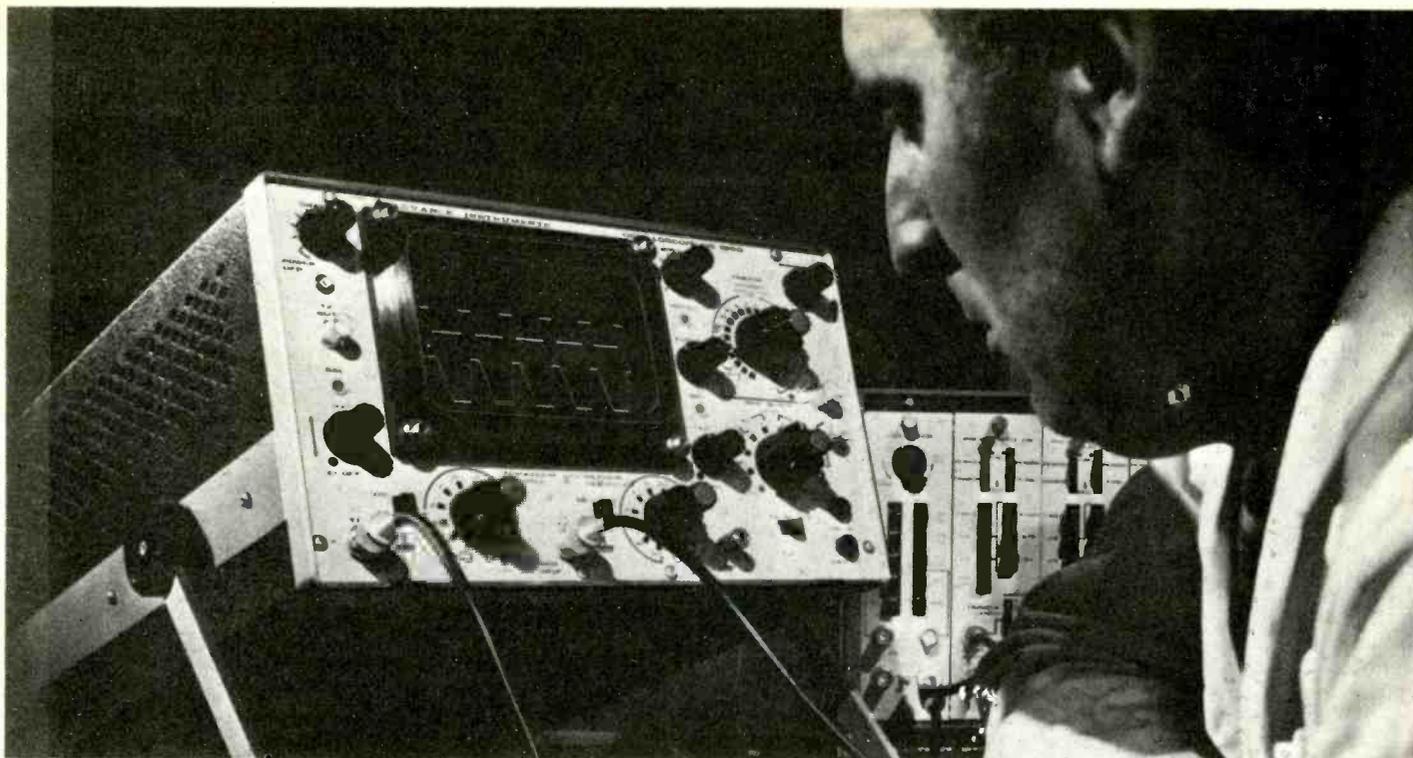
Beaverton House, P.O. Box 69,  
Harpenden, Herts.

Tel: Harpenden 61251 Telex: 25559

Northern Region Office:

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Comprehensive trigger facilities with T.V. sync separator.

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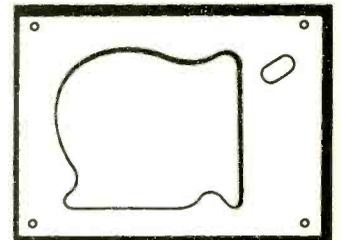
Raynham Road,  
Bishop's Stortford, Herts.

Telephone:  
Bishop's Stortford (0279) 55155  
Telex: 81510

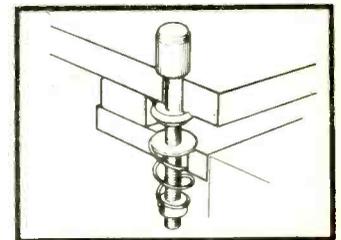


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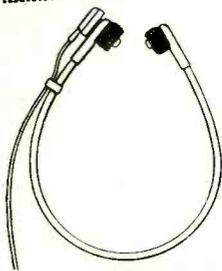
*Four-point spring suspension adjustable for height and damping protects the motor board from acoustic feedback and external vibration.*

# SME

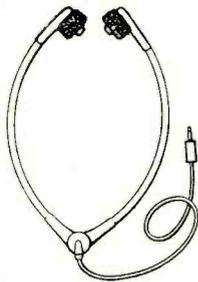
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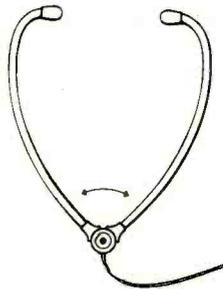
... on accessories for dictating machines, tape recorders, tele-communications and electro acoustic equipment, etc.



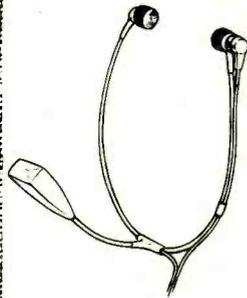
STETOCLIP JUNIOR 60 HEADSET



STETOCLIP LIGHTWEIGHT HEADSET



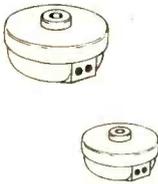
STETOCLIP SENIOR HEADSET



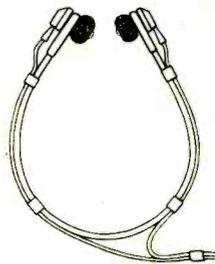
STETOMIKE BOOM MICROPHONE HEADSET



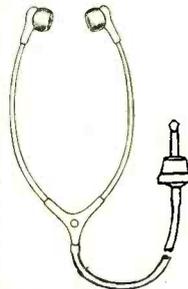
PLASTIC & NYLON EARHANGERS



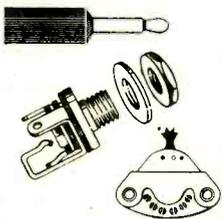
STANDARD & SUB-MINOR EARPHONES



STEREOCLIP HEADSET



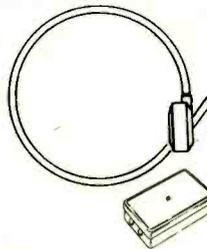
STETOTUBE HEADSET & SOUNDPLUG FOR HOSPITALS



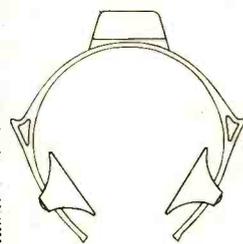
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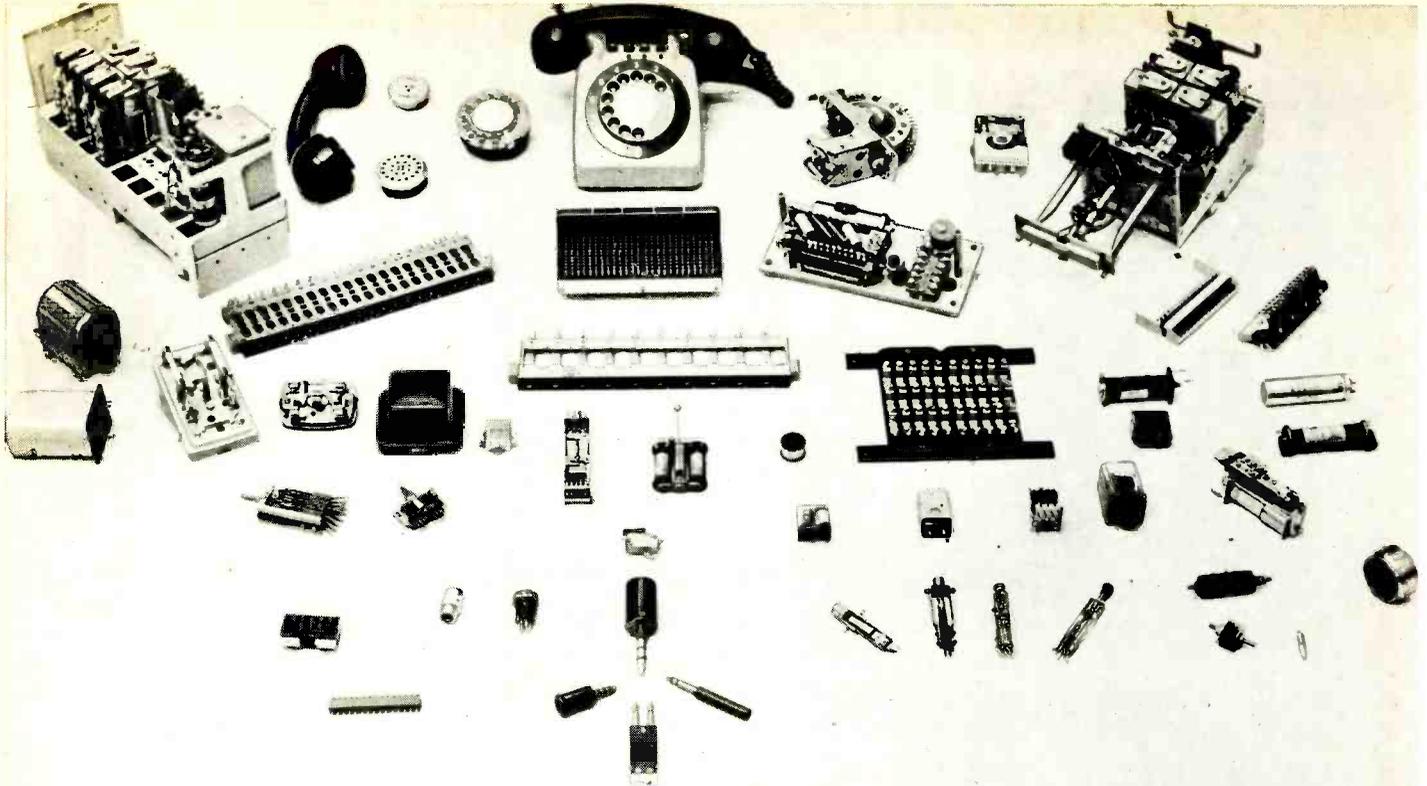
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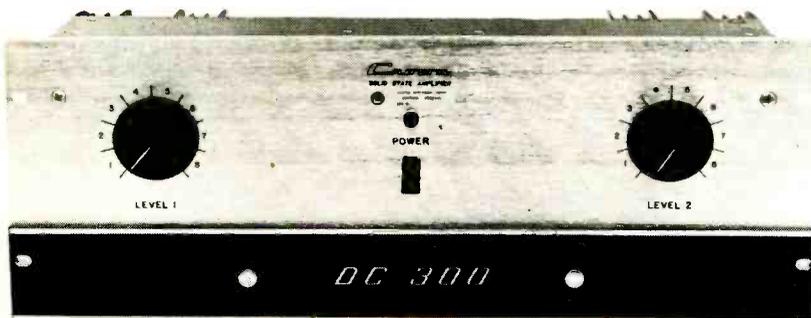
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Tel. 01-203 2814 Cables: Telwireco, London, N.W.9.

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## DC300

## DUAL-CHANNEL POWER AMPLIFIER



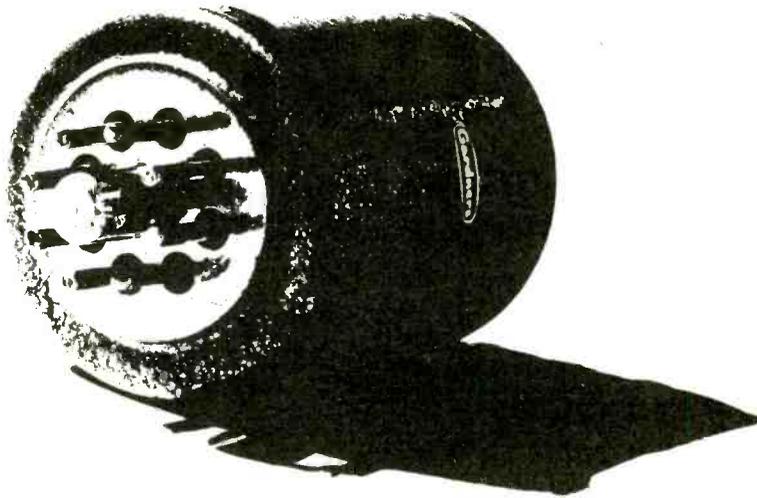
Frequency Response	$\pm 0.1\text{db}$ Zero-20KHz at 1 watt into 8 ohms, $\pm 0.6\text{db}$ Zero-100K Hz.
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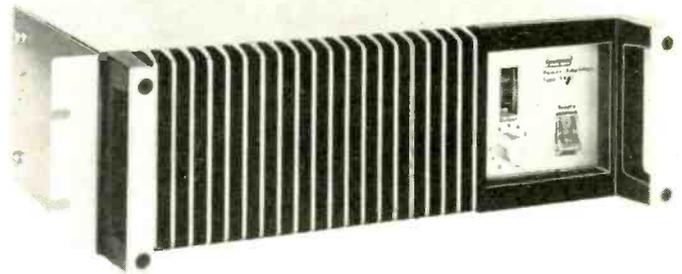
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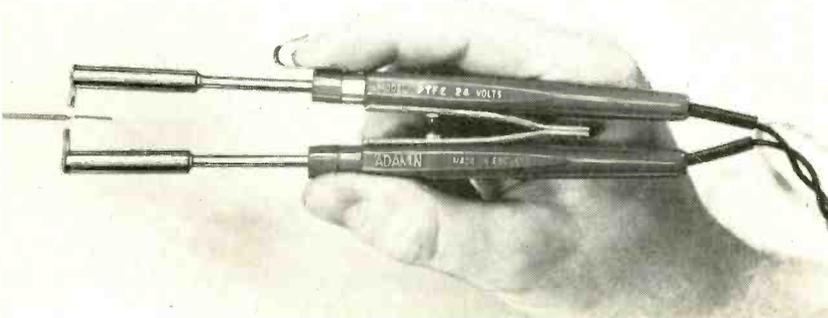
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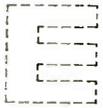
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## 'Toa' PA-III Meeting Amplifier

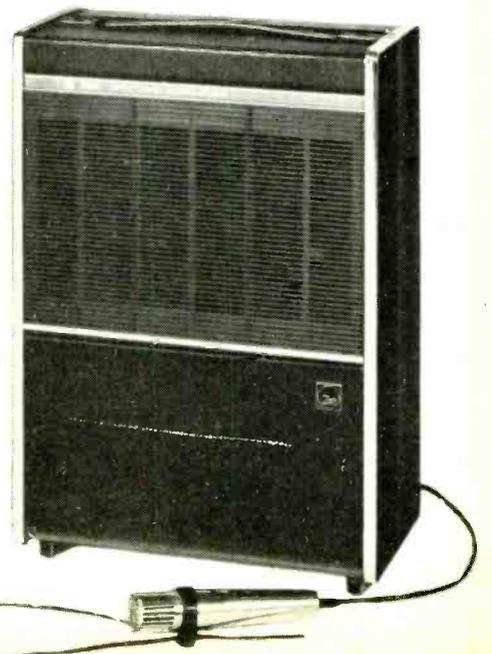
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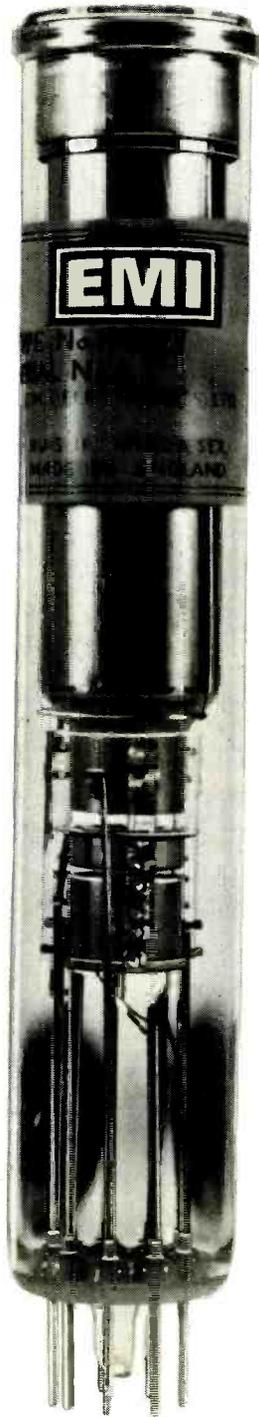
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\* 5LB

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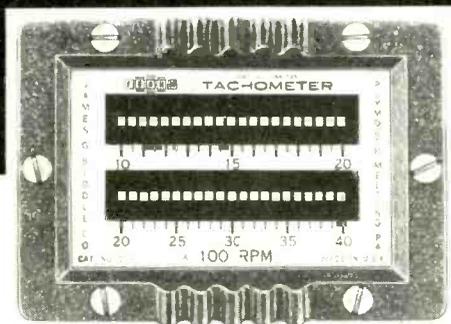
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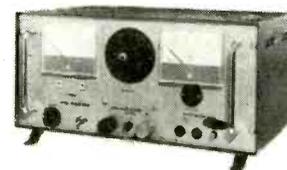
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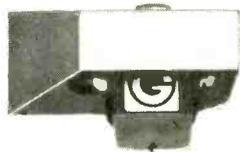
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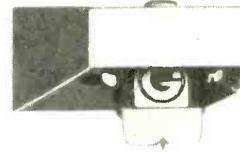
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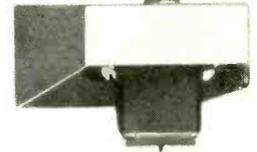
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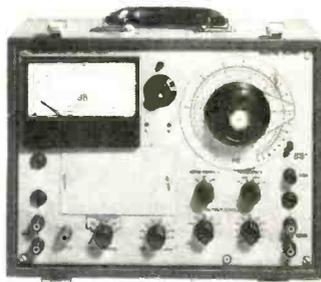


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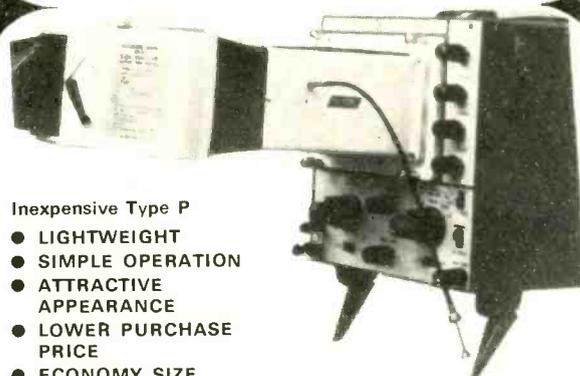
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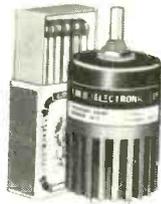
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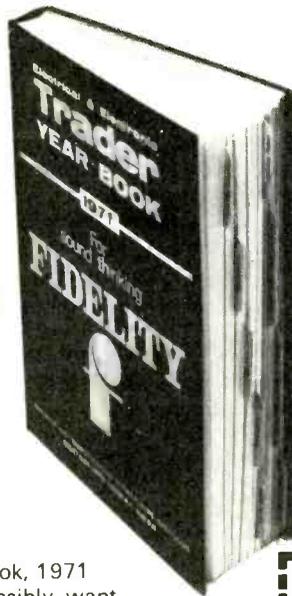
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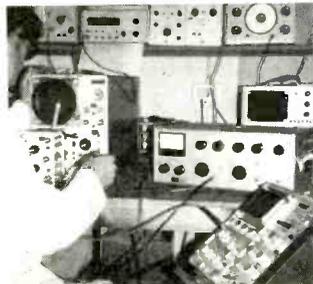
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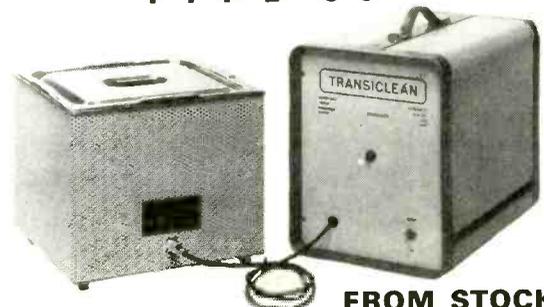
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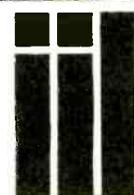
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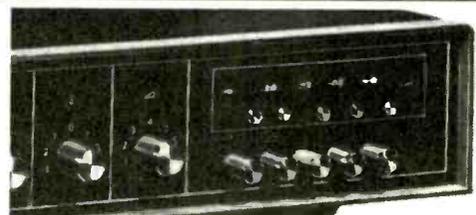
### 840A amplifier design features

Swing-up lid enables all input and output connections to be reached instantly and easily from interior of cabinet. Audio connections are via DIN sockets (plugs supplied). Combined speaker muting switch and headphone socket on front panel. Silicon transistors throughout. Full complementary output stages with current limiting circuitry. Rotary controls fitted with dual wipers and lubricated tracks for long life and silent action. 18 gauge plated steel chassis. Unconditionally stable. For shelf or cabinet mounting.

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Facilities—Swing-up lid for instant access to connections, permitting cabinet to stand flush to wall; also space to take Englefield tuner if required.

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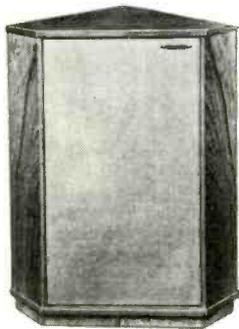
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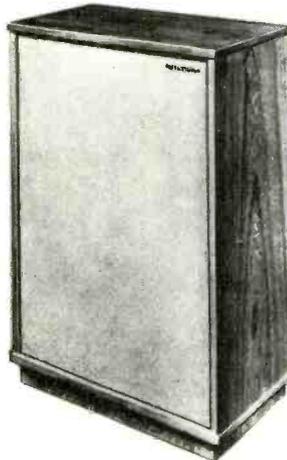
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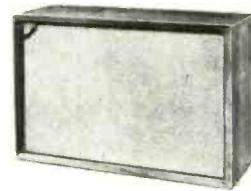
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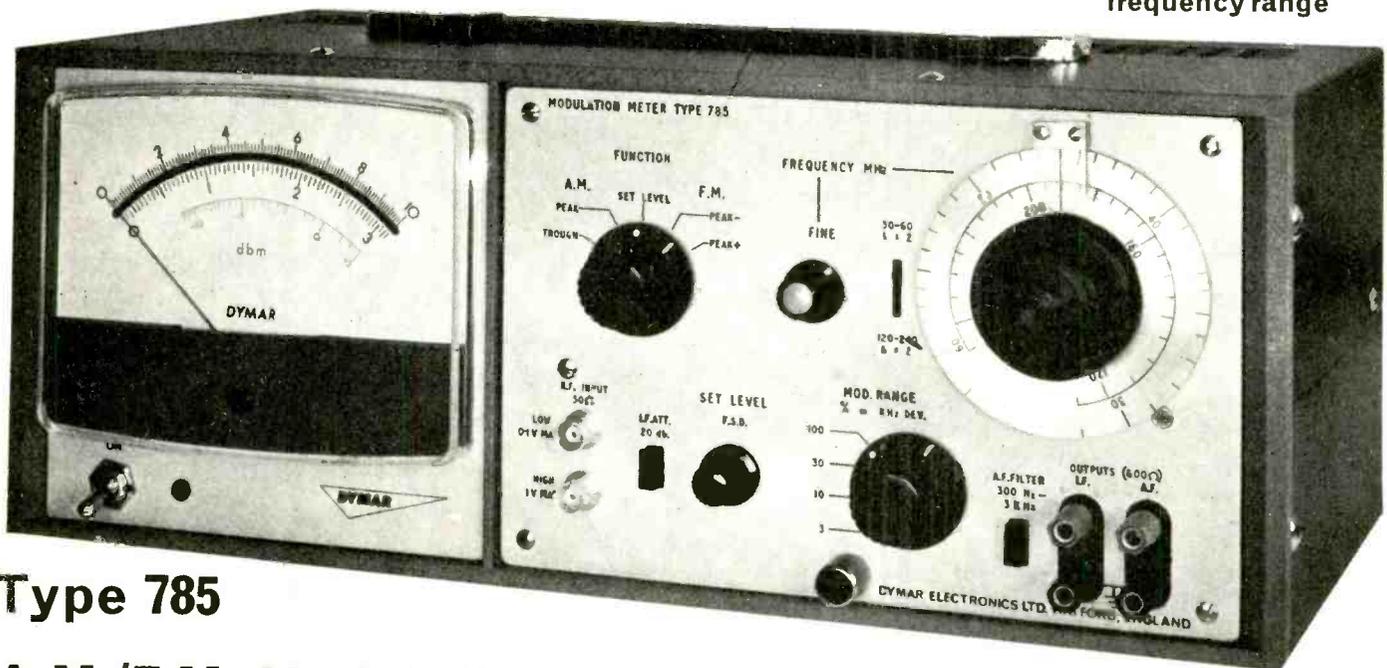
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After each unit has been passed through visual inspection it is given a full specification check in the test laboratories prior to going on 'soak test'.

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Why have we incorporated this in to our established test procedures, which have always been considered rigorous?

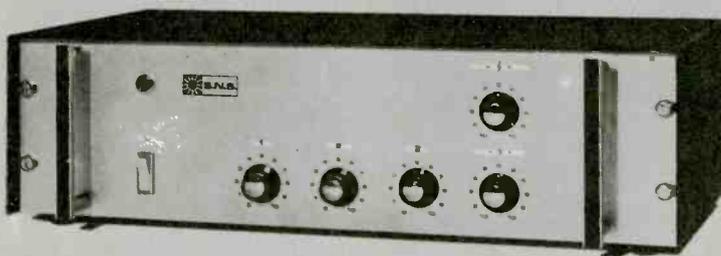
Simply as an added safeguard to ensure that our customers throughout the world receive equipment with the highest possible reliability.

It's no coincidence that S.N.S. equipment has a reputation for quality . . . it's been designed that way.

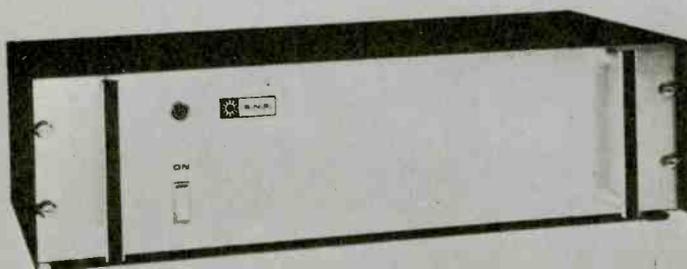
## range of sound equipment

S.N.S. Communications Ltd., 851 Ringwood Road, West Howe,  
Bournemouth, England, BH11 8LN Telephone: Northbourne 5331 (STD 02016)

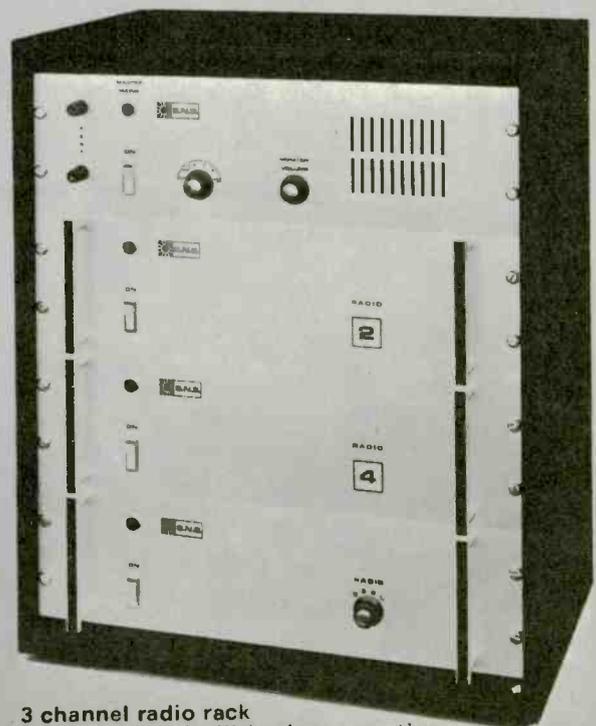
# Tested for 72 LONG hours



40 watt multi input amplifier Type PA40



100 watt single input amplifier Type CD100



3 channel radio rack  
with monitor facilities incorporating  
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# News of the Decade

**Capacitance boxes** available from 20pf.-140uf. Accuracies up to 0.05%

## Inductance

Air space 1mH-1H  
Accuracy 5%

## Resistance Boxes

from 0.1-10 MΩ  
Average accuracy 0.1%

Resistance Elements suitable for use up to 1MHz.



### High Dissipation Resistance Boxes

*Five Decade*

HD1	0 to 1,111,100 ohms by 10 ohm steps	1% Tolerance	£45.00
HD5	0 to 1,111,100 ohms by 10 ohm steps	5% Tolerance	£35.00
HD1/L	0 to 111,110 ohms by 0.1 ohm increments	1% Tolerance	£47.00

### Jay-Jay Inductance Boxes

Cat. Ref.			
L1	3 Decade 1 mH to 1 Henry		£37.00
L2	2 Decade 1 mH to 100 mH		£26.00
L3	2 Decade 10 mH to 1 Henry		£29.45

### Jay-Jay "Point One" Resistance Boxes

Cat. Ref.	Average Accuracy 0.1%		
	<i>Four Decade</i>		
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R4	0 to 11,110 ohms by 1 ohm steps		£24.50
R3	0 to 1,111 ohms by 1/10 ohm steps		£25.00
	<i>Five Decade</i>		
R11	0 to 10 meg ohms by 100 ohm steps		£34.50
R7	0 to 1,111,100 ohms by 10 ohm steps		£29.25
R9	0 to 111,110 ohms by 1 ohm steps		£29.75
R10	0 to 11,111 ohms by 1/10 steps		£30.25
	<i>Six Decade</i>		
R20	To 1,111,110 by 1 ohm steps		£36.00
R21	02 111,111 by 0.1 ohm steps		£36.50
R22	0 to 11,111.1 by 0.01 ohm steps		£37.00

### Jay-Jay Junior Decade Resistance Boxes

Cat. Ref.	Average Accuracy 0.4%		
	<i>Six Decade</i>		
J60	Range 0 to 1,111,100 ohm by 1 ohm steps		£18.90
	<i>Five Decade</i>		
J1	Range 0 to 1,111,100 by 10 ohm steps		£15.60
J2	Range 0 to 111,110 by 1 ohm steps		£15.40
	<i>Four Decade</i>		
J3	Range 0 to 111,100 by 10 ohm steps		£12.40
J4	Range 0 to 11,110 by 1 ohm steps		£12.20
	<i>Three Decade</i>		
J5	Range 0 to 11,100 by 10 ohm steps		£9.95
J6	Range 0 to 1,110 by 1 ohm steps		£9.90

### Jay-Jay Capacitance Boxes

Cat. Ref.			
	<i>Three Decade Model</i>		
C3	100 pf. to 0.111 mfd.	1% Tolerance	£22.00
PC3	100 pf. to 0.111 mfd.	½% Tolerance	£31.50
	<i>Four Decade Model</i>		
C4	100 pf. to 1.111 mfd.	1% Tolerance	£35.00
PC4	100 pf. to 1.111 mfd.	½% Tolerance	£48.00
	<i>Switched Capacitance Boxes</i>		
C100	1 mfd. to 100 mfd.	5% Tolerance	£65.00
C140	1 mfd. to 140 mfd.	5% Tolerance	£75.00
C60	0.1 mfd. to 61 mfd.	5% Tolerance	£59.00
C60P	0.1 mfd. to 61 mfd.	1% Tolerance	£110.00
	<i>Air Spaced Capacitors</i>		
VC1	10 to 160 pf.	1% Tolerance	£12.00
	<i>Variable Capacitors</i>		
VC2	20 to 1130 pf.	1% Tolerance	£22.00
VC4	50 pf. to 0.1114 mfd.	1% Tolerance	£28.75
VC5	50 pf. to 1.1114 mfd.	1% Tolerance	£43.00
	<i>Precision Air Spaced Capacitors</i>		
PVC1	5 to 200 pf.	½% Tolerance	£36.00
	<i>Precision Variable Capacitor</i>		
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	<i>Precision Incremental Capacity</i>		
PVC4	± 0.5 pf. and ± 5 pf.	1% Tolerance	£36.00
	<i>Precision Variable Capacitor</i>		
SVC5	50 pf. to 1.1114 mfd. ± 0.05% tolerance—major decade		£230.00

### Jay-Jay Junior Decade Capacitance Boxes

Cat. Ref.	Accuracy 1%		
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JC2	3 Decade 30 pf. to 10, 140 pf.		£13.70

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**R.E.C.M.F.**  
STAND NO.  
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# The Dolby 360 Series

Nearly a thousand of these new units are already in use.



Each Series 360 unit is only 1 $\frac{3}{4}$  inches (44 mm) high. 16 channels therefore require only 28 inches of rack space.

## Full compatibility with the A301

Models 360 and 361 are single-channel A-type (professional) noise reduction units which process signals identically to the two-channel A301. The new units are small in size and are designed for simplified installation and use of the Dolby System with 16-track recorders. The cost of the 360 series is somewhat less than that of the A301 for an equivalent number of channels.

## Automatic record/play changeover in the 361

The Model 360 is a single-channel noise reduction processor unit. The Model 361 is identical to the 360 in size and appearance, but contains facilities for automatic record/play changeover controlled from the recorder. In the new series, the operating mode is set and clearly displayed by illuminated push-button switches.

### Internal oscillator

An internal "Dolby Tone" oscillator is provided for establishing correct operating levels. The characteristic modulation of the tone also identifies Dolby-processed tapes. All oscillators in a multi-track installation can be controlled by a single switch.

### High stability

The circuit is highly stable and does not require routine adjustment. A removable front panel allows input and output levels to be adjusted from the front of each unit. The panel also provides access to relays and the noise reduction module.

### Single-module design

The noise reduction circuitry is contained in a single module which can be purchased separately. Should failure ever occur, plug-in substitution will restore operation of the system in seconds with no adjustments necessary.

Prices, delivery information and complete specifications are available from



## DOLBY LABORATORIES INC

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(212) 243-2525 cables: Dolbylabs New York

UK and International  
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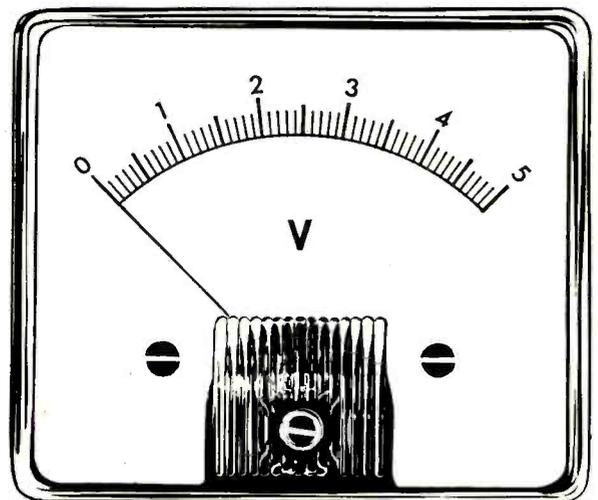
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**METER PROBLEMS?**



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# Sansui control amplifiers. Whenever you're ready.

Whenever you're ready for the added professionalism of a control amplifier, Sansui is also ready. With a complete line of quality units in every power and price range. Each a distinguished example of Sansui's dedication to stereophonic perfection.

If you're ready for the most complete, the most powerful such amplifier, then the 180 watt solid state AU-999 is for you. The undisputed champion in the field, it offers the most advanced circuitry yet developed, boasts a 10 to 30,000Hz power bandwidth and keeps distortion of any type to 0.4% or less. The AU-999 also offers separately usable pre- and power amplifiers, a Triple Tone Control circuit, and the capability of handling three sets of speaker systems and two tape decks.

For those more inclined toward the middle power ranges, there's the 100 watt AU-666 and 85 watt AU-555A, both striking examples of engineering excellence. The AU-666 has a wide 10 to 40,000Hz power bandwidth and limits distortion to a low 0.5%, while the AU-555A has a 20 to 40,000Hz power bandwidth and the same low distortion figure. Both are complete with separately usable pre- and

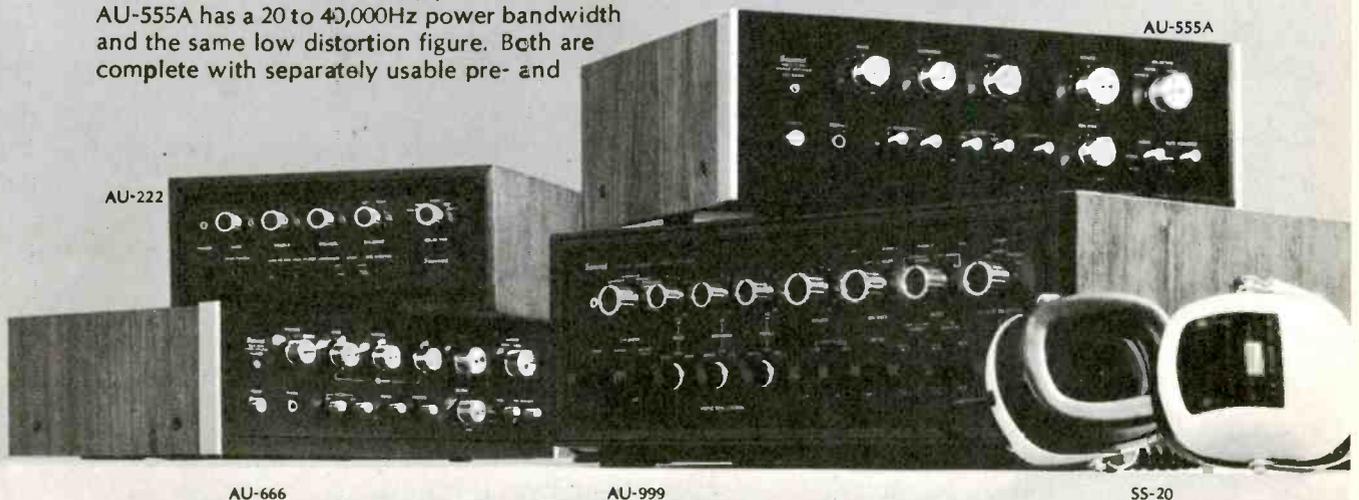
power amplifier sections and Sansui's unique Triple Tone Control circuit.

And in compacts, nothing quite approaches Sansui's 46 watt AU-222 for completeness and versatility. Endowed with a 20 to 20,000Hz power bandwidth, it minimizes distortion at 0.8% or less, and offers no fewer than six inputs — more than any other amplifier of its size.

Whichever you choose, enjoy it more fully with Sansui's advanced SS-20 stereo headphone set, a 2-way 4-speaker unit complete with separate controls for tonal quality and volume. No other stereo headphone set is geared so perfectly to exploiting Sansui control amplifiers.

So whenever you're ready for the undeniable advantages of such an amplifier, see your nearest authorized Sansui dealer. He's ready with a full selection.

*Sansui*



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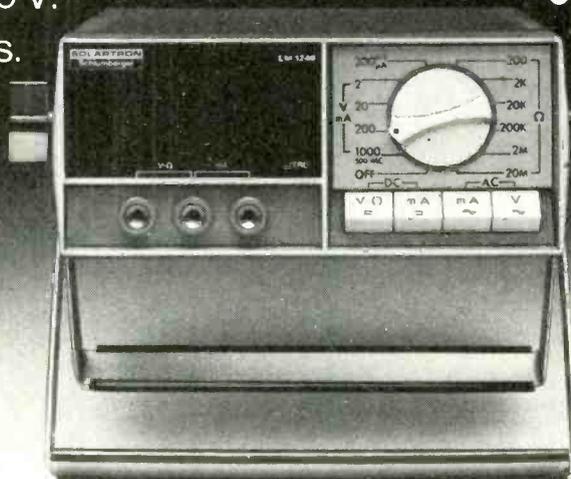
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So if when you order your LM 1240 direct from this ad., you are not completely satisfied that it does all we say of it, we'll take it back within 14 days and return your money in full.

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# TRANSFORMERS

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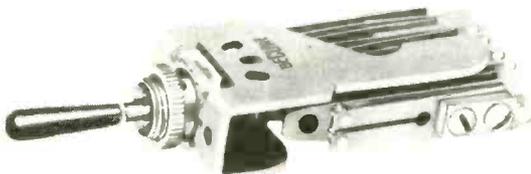
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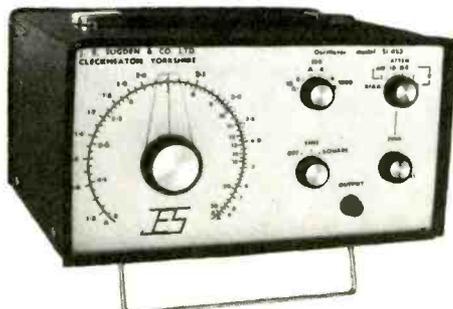
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#### SPECIAL FEATURES:

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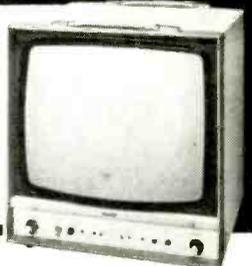
# prowest *for* PICTURE MONITORS

Brief details of the wide range of all-silicon solid-state monitors are shown below. Please contact us for full information.

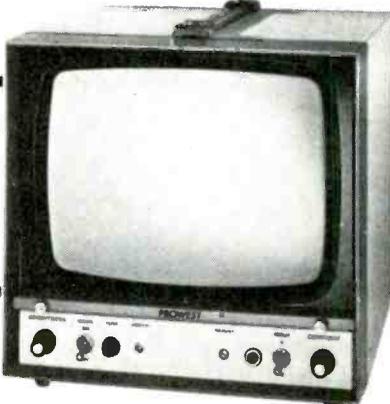
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14"  
19"  
20"  
(20"illuminantD)

**Series 1A**

High quality monochrome monitors suitable for use either in T.V. studios or laboratory applications demanding the highest possible performance. Facilities include electrical centering, modular circuit design and separately stabilised E.H.T. These together with exceptional stability and high brightness capability are some of the notable features of this range.

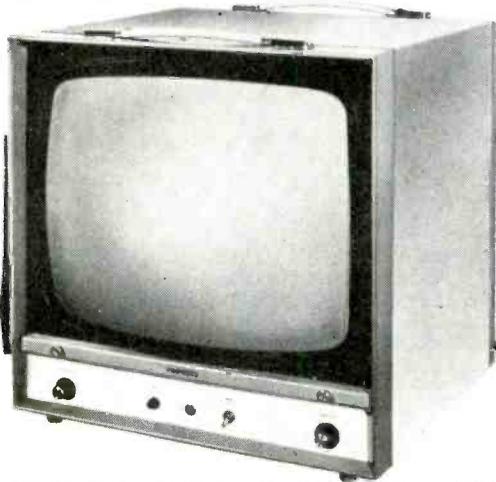


Designed for a wider range of less demanding applications and offering an extended choice of screen sizes. Ideally suited for general studio uses, high performance industrial systems and data display. Many optional features are available including remote control facilities and special c.r.t. phosphors.



**Screen Sizes**  
11"  
14"  
19"  
24"

**Series 3A**



The economy range of monochrome monitors incorporating the already familiar Prowest high standards of construction and design. The attractive price, smart appearance and rugged construction make these displays suitable for a wide range of office and industrial environments.

**Screen Sizes**  
12"  
19"

**Series 5A**

**Screen Size**  
22"

**Series 7A**

A keenly priced precision colour monitor in the grade 1 class using the latest 22" 4 : 3 aspect ratio shadow mask c.r.t. Excellent stability, remote operation of certain user controls, integral plug-in PAL or N.T.S.C. decoder, natural convection cooling and front panel purity control are some of the notable features.




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# High Fidelity Equipment

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These headphones are designed for high fidelity reproduction of music. They feature a wide frequency response and excellent sound reproduction. The ear cups are large and comfortable, and the headband is padded for long listening sessions.



These headphones are designed for high fidelity reproduction of music. They feature a wide frequency response and excellent sound reproduction. The ear cups are large and comfortable, and the headband is padded for long listening sessions.



These headphones are designed for high fidelity reproduction of music. They feature a wide frequency response and excellent sound reproduction. The ear cups are large and comfortable, and the headband is padded for long listening sessions.



# The Eagle Annual.

Sorry, no Dan Dare, Digby or P.C. 49. Because this is the new Eagle annual catalogue. And it's packed with interesting things. Like the new TSA 151 stereo amplifier: it uses a new block construction silicon output device for absolute reliability. It's got low noise silicon transistors throughout. Its output is 15 Watts per channel. That's 15 Watts RMS, not an exaggerated figure for maximum music power.

The price? A very reasonable £36.

And for people who like to listen to stereo undisturbed, we've got the new SE 100 headphones.

Dual cone transducers are used throughout, and to keep the weight down, the independent volume controls are

mounted on a separate unit with a pocket clip. £16.00.

Every item in the annual has been specified or selected by Gerry Adler. Eagle is Gerry's baby, and he's very fussy about what goes out under the Eagle banner. He gets very twitchy at the thought of a duff diode. A bit like the Mekon in fact.

But he does it for a reason.

He believes that if the first Eagle product you buy is O.K., you'll come back for more.

That's what's made Eagle a success.

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we don't stand still.

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## Headphones/Accessories



**SE 78 Studio Stereo Headphones**  
This studio quality headset offers a really extensive frequency response (20 to 20,000 Hz) by means of dome acoustic chambers and specially designed tweeters. Separate adjustable attenuators are also incorporated in each dome. Complete with stereo jack plug and cable - a first-class instrument.

Frequency Range: 20-20,000 Hz. Impedance: 8 ohms per channel. Matching Impedance: 8-16 ohms.



**SE 80 Studio Stereo Headphones**  
Based on an entirely new open baffle concept the SE 80 Stereo Headphones have been developed to bring a new standard of comfort to professionals who have to wear a headset for very long sessions. Recording Engineers, Monitor Units, Broadcasters, D.J.'s, etc. Individual Slider tone controls complete a list of features which include the use of fine leathers for padding.

Frequency Range: 20-20,000 Hz. Impedance: 8 ohms per channel. Matching Impedance: 8-16 ohms per channel.



**HMA 309 Headphone and Boom Microphone**  
Ideally suited to the language laboratory, studio or other professional user, this new headphone and boom microphone are the result of combined UK and Canadian research. A new noise cancelling microphone is inset in a specially moulded boom which carries the microphone cable internally.

Frequency Range: 20-14,000 Hz. Impedance: 8 ohms. Matching Impedance: 8-16 ohms. Cord length: 1.8m. Dimensions: 125mm x 100mm x 45mm.



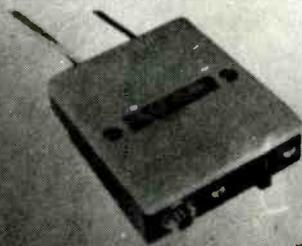
**SE 100 Professional Stereo Headphones**  
The SE 100 represents the ultimate development of classic design. Built in differences that are audible include tone corrected dual cone speakers and damped reflex baffles. To keep headset weight to an absolute minimum the independent volume controls are mounted in a separate unit. Leather padding, a 5 metre coiled lead and a moulded stereo jack plug finalise a luxury specification.

Frequency Range: 20-21,000 Hz. Impedance: 8 ohms. Matching Impedance: 8-16 ohms.



**JB 3 Stereo Headphone Junction Box**  
This simple, inexpensive unit is essential to stereo headphone users. It connects to amplifier and loudspeakers, giving attenuated stereo headphone output, with a three-position switch to give headphones only, speakers only, or both together.

Input: Suitable for connection to any amplifier up to 20 watts output. Output: 50 mV. Size: 74 x 50 x 28 mm.

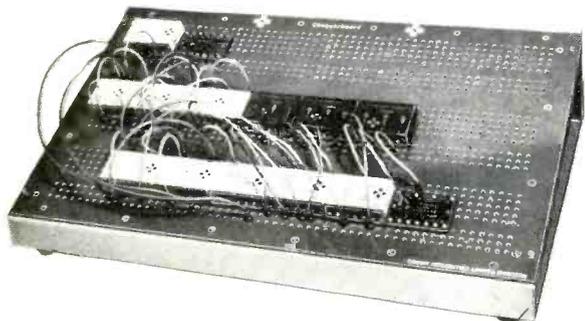


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Input: Suitable for connection to any amplifier up to 20 watts output. Output: 50 mV. Size: 74 x 50 x 28 mm.

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Chequerboards provide an easy plug-in facility for all integrated circuit modules. Interconnection is via plug-in cords for easy non-soldering interchange of all modules.

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We shall be glad to send you our 15 page fully illustrated brochure, telling you all about CHEQUERBOARD PATCHBOARDS — ACCESSORIES — POWER UNITS — EDUCATIONAL KITS, etc.

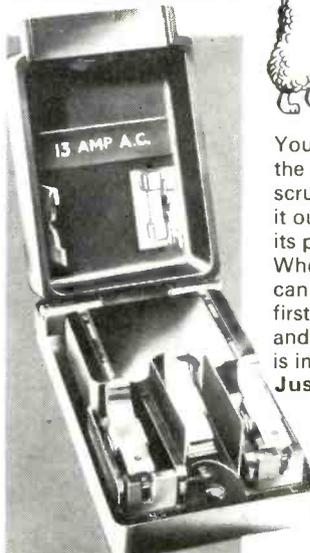
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You choose electrical equipment in much the same way you choose a dog: you scrutinise and examine carefully, and take it out for a trial run. And you ask to see its pedigree.

When you handle a Rendar product, you can see at once that it's a better breed: first class materials; precision machinery and assembly; sound design. Its pedigree is impeccable.

**Just 1 winner from the Rendar kennel**

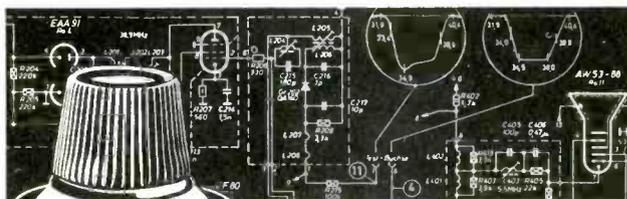
The Rendar Safeloc saves time and saves life. There's no need to fit a plug for testing — just connect the apparatus direct. And there's no danger of shocks — no current can pass until the lid is closed. Rendar pioneered this concept, and introduced the "Safeloc" to the British market over 12 years ago. It's indispensable on testing lines, and for all kinds of electrical demonstrations. Double Safeloc available for 3-phase applications.



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INSTRUMENTS LTD

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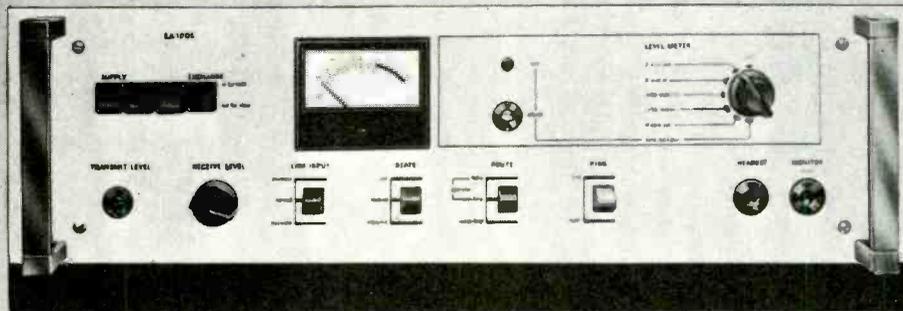
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# Racal bridges a gap in communications

with this new radiotelephone terminal  
type LA 1005



In quantity production after successful world-wide acceptance trials, the Racal LA 1005 provides exceptional system flexibility and operational facilities at low cost. Extremely compact, it is suitable for Post and Telegraph Administrations' circuits and ideal for military or civil operation.

- 4 wire radiotelephone to 2 wire or 4 wire telephone circuit.
- Automatic voice operated switching. ● Fully transistorised.
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FLUX: 128,000 MAXWELLS  
IMPEDANCE: 15 or 4-8 OHMS  
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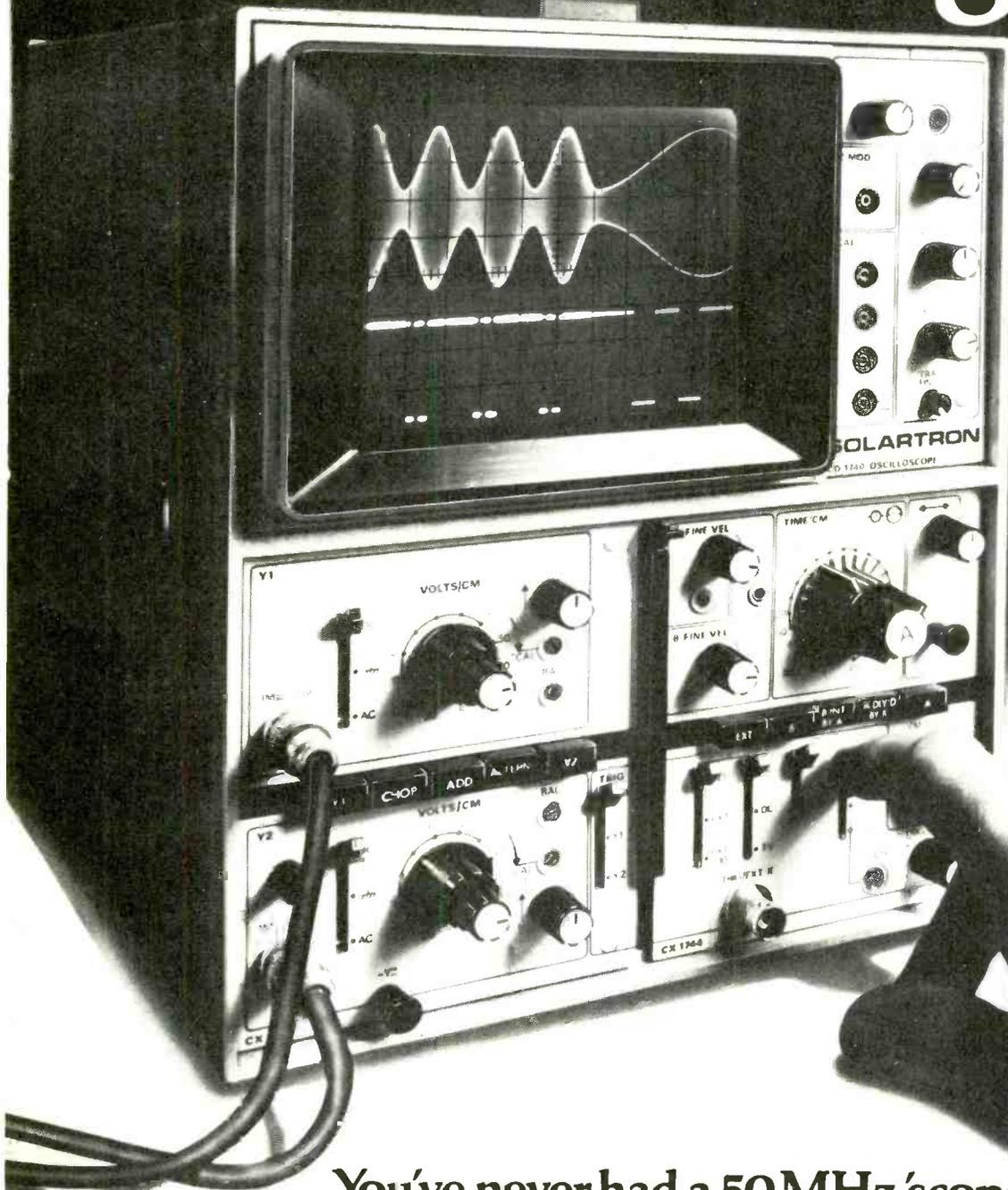
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The push-buttons and lever switches are where you want them, when you want them, on a compact:  $8\frac{3}{4}'' \times 10\frac{1}{2}''$  front panel.

Take the trigger controls for example. With the five lever switches up, you have fully automatic triggering over the full bandwidth with a bright base line under no signal conditions.

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Further, there's a sure-fire trace finder button which brings a free-run brightened trace on screen—whatever the control settings.

But handling apart, the CD1740 packs real performance.

The timebase push-buttons give you 10 sweep modes including delayed and delayed gating with the capability of closely inspecting complex wave forms without jitter.

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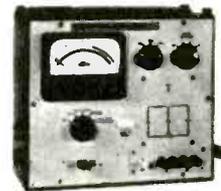


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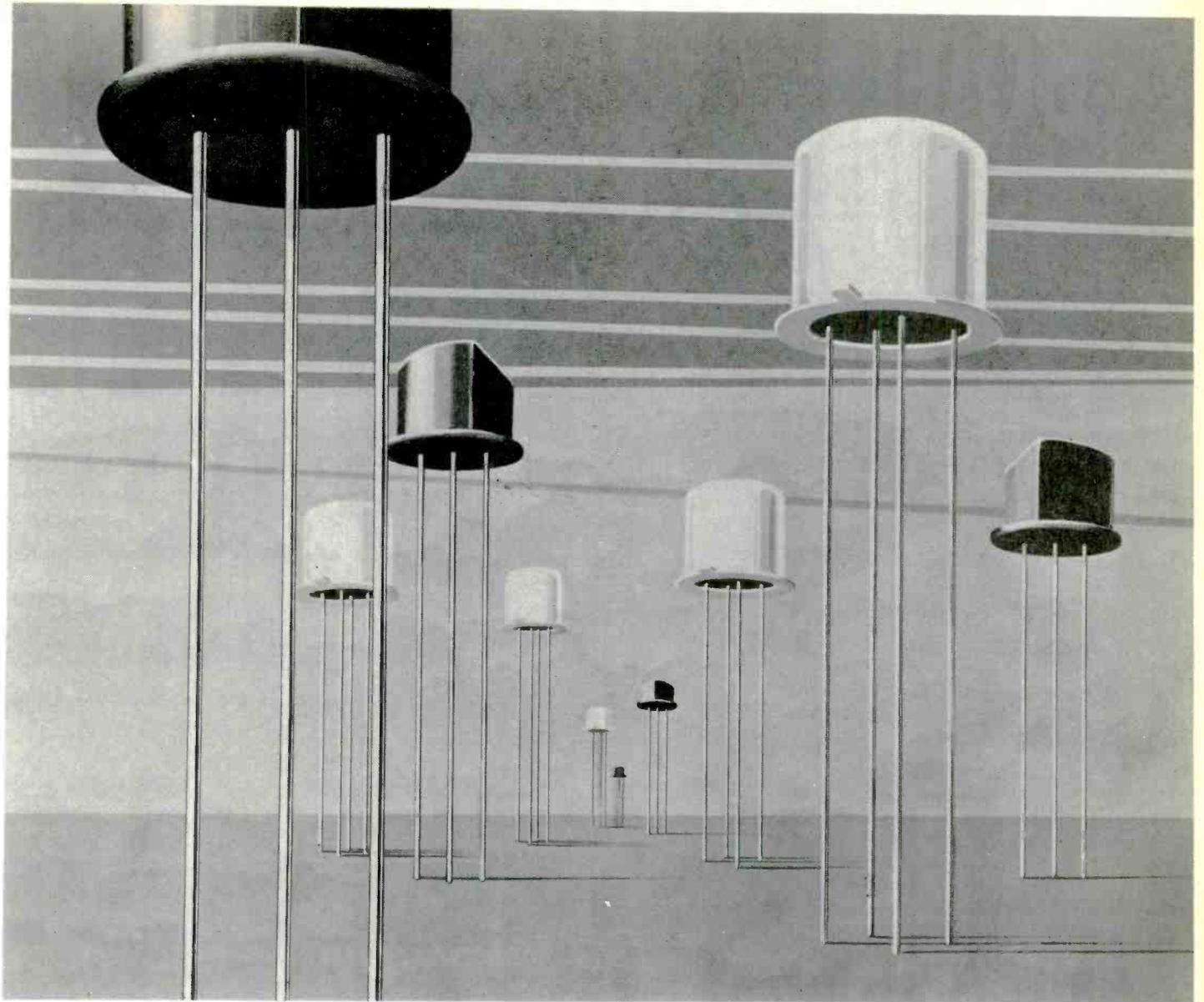
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The output can be switched to provide either 0-30V, 0.5A or 0-15V, 1A with full current capability over the entire voltage range. Precise levels of voltage and current are monitored by a large clear scale meter. The power supply is protected from overload by current limiting circuitry.

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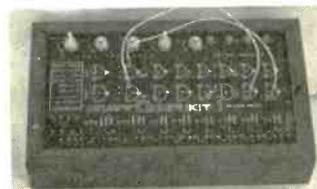
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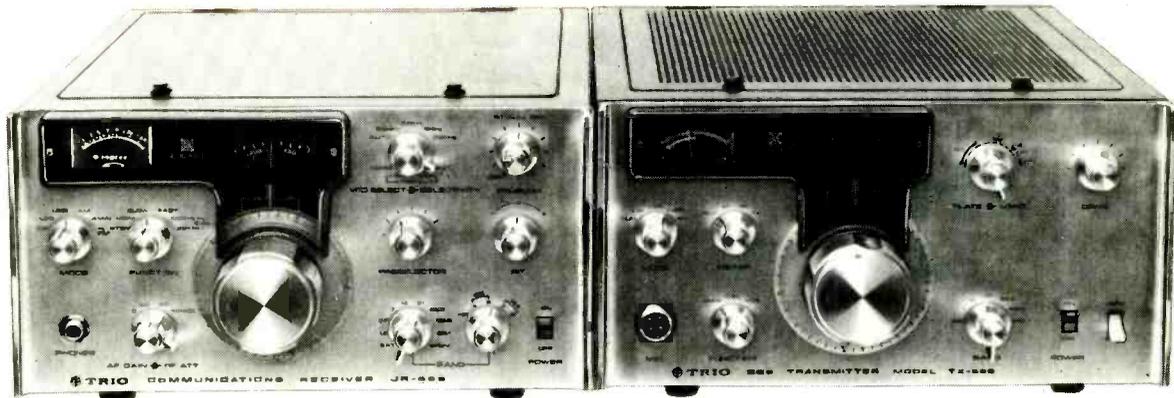
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ALL BAND COMMUNICATIONS RECEIVER JR-599

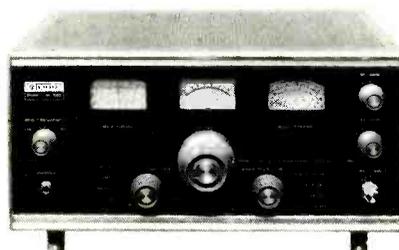
ALL BAND SSB TRANSMITTER TX-599

## TRIO IS FULLY-EQUIPPED FOR FULL-CYCLE COMMUNICATIONS

TRIO's JR-599 communications receiver brings the highest-type, professional, all-bands potential to amateur bands on an allocated 1.8 to 29.7 MHz frequency range, 50 and 144 MHz bands and WWV's 10 MHz standard signal. A receiver frequency readable to the nearest 500 Hz is guaranteed due to precision type double gear mechanism and variable capacitor with linear characteristic for main tuning dial of a 25 kHz band at one full turn. The all-band SSB TX-599 transmitter matches the JR-599 with its wide-spread IC and FET network. All HF bands are covered with its single switch mode on LSB, USB, AM and CW positions. All of TRIO's equipment—or equipment combinations—is designed to provide entirely full-cycle communications capability.

### SP-5D COMMUNICATIONS SPEAKER

- Communications Speaker which has been designed for use with the 9R-59DS.
- Dimensions: 3-9/16" (W), 7-1/8" (H), 5-3/16" (D).



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at Rotel hi-fi  
equipment  
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clearly in your  
head...**

# The Sound

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But, before we even give you a sight of it, we want you to hear something first. For example, that Rotel, still a fairly new name on the hi-fi scene, is building a reputation for the best value-for-money equipment available. That Rotel sports some technical features surprisingly advanced for its price range. That Rotel equipment is being handled and serviced in the UK by The Rank Organisation. And that the Rotel sound sounds right to even the keenest ear.

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**But whichever you do, do it soon, d'you hear?**



# ROTEL

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Rank Audio Visual Ltd  
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Middlesex Telephone 01-568 9222

#### **RA 310 Stereo Amplifier**

Power output: 15 watts RMS per channel into 8 ohm.  
Frequency response: 20-30,000 Hz + 0 - 1.5.  
Input sensitivity: Mag 3 mV, Aux 200 mV, X'tal 100 mV,  
Tape Monitor 300 mV.  
Hum and noise: Phone 60 dB, Aux 65 dB, Tuner 65 dB.  
Separate bass and treble control. Tape monitor  
switch. Provision for two pairs of speakers (perfect for  
parties, or upstairs-downstairs listening).

**£42.50** (rec. retail price).

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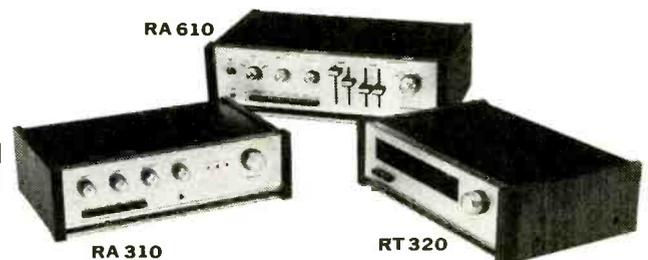
Sensitivity: 2.5  $\mu$ V.  
Signal to noise: 60 dB.  
Stereo separation: 35 dB.  
Price complete with stereo decoder

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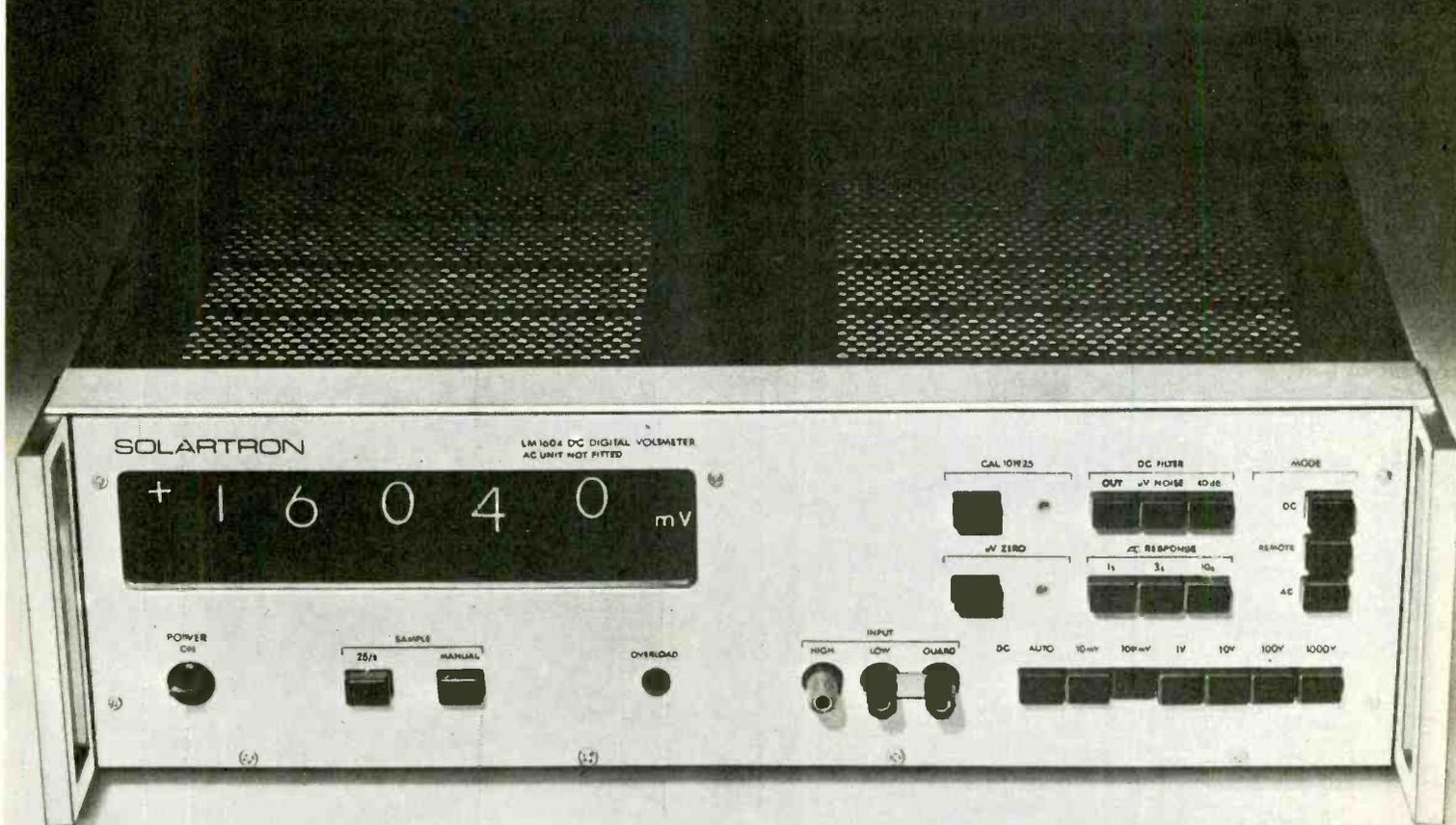
#### **RA 610 Stereo Amplifier**

Power output: 30 watts RMS per channel into 8 ohm.  
Frequency response: 20-50,000 Hz + 0 - 1.5.  
Input sensitivity: Mag 3 mV, Aux 200 mV, X'tal 100 mV,  
Tape Monitor 300 mV.  
Hum and noise: Phono 60 dB, Aux 70 dB, Tape 70 dB,  
Tuner 70 dB.  
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Europe's engineers have been acclaiming Solartron's LM1604 as the most magnificent all-round performance, value-for-money DVM on the market.

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Now armed forces in Europe are using it.

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delay 50 ns–1 sec. width 50 ns–1 sec.
- \* Gating, external trigger and manual one-shot facilities
- \* *True double pulse*—two channels each with independent delay, width and amplitude
- \* Pulse advance capability to +1 sec.
- \* Super-portable: only 3½" x 9¼" x 11", 7½ lb  
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- \* Internal channel mixing facilities



## it sells itself!

More details of the new PG-71 please.

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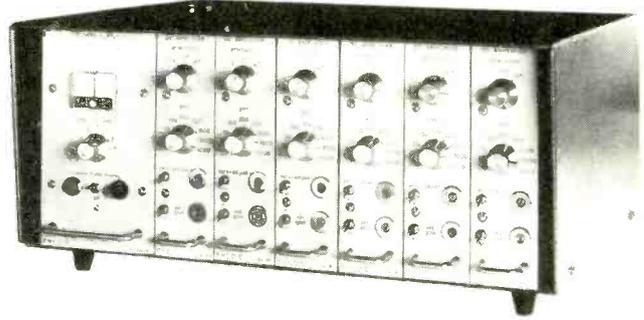


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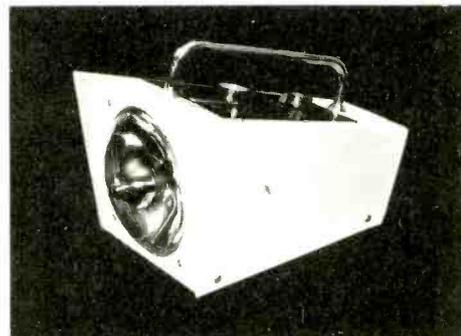
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Compatible modules and cards ensure  
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## XENON STROBOSCOPE



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The instrument is of modern appearance, small, light in weight, convenient to use and portable. A wide range of flashing rates is covered by the large accurately calibrated dial, allowing operation at low frequencies for stroboscopic photographic experiments and at high speeds for observation of rapidly rotating or reciprocating phenomena.

The external triggering facility permits single shot operation by an external closing contact and also provides a synchronising input for high and low speed repetitive phenomena which might otherwise be difficult to maintain in exact phase.

Light source.

High intensity Xenon tube mounted in a parabolic reflector.

Flashing rate.

1-250 flashes/second in 3 ranges.

Frequency accuracy.

Typically ± 2% of each full scale.

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(a) by internal oscillator  
(b) by external closing contacts.

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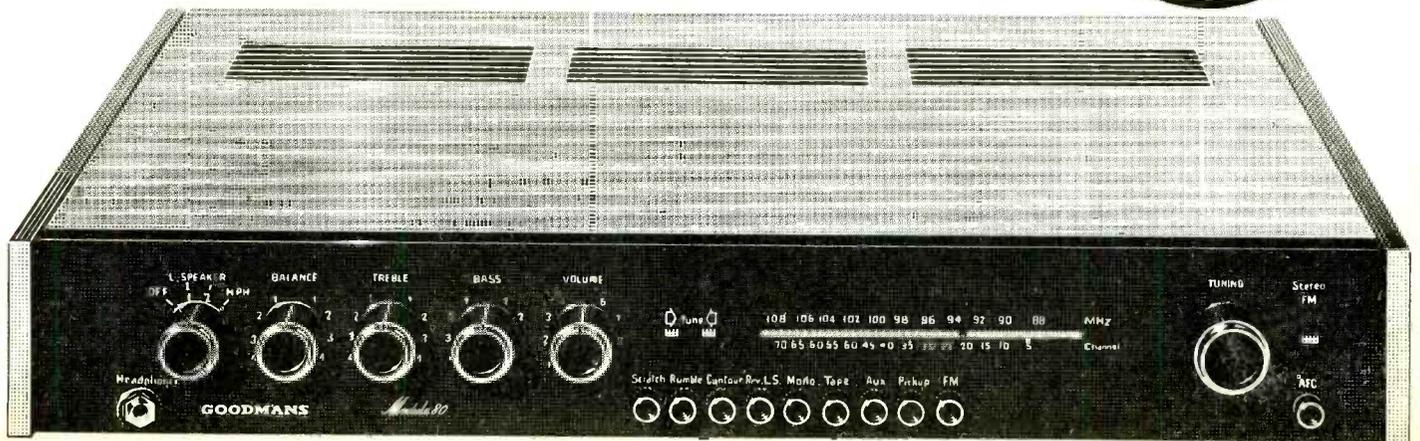
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Europe's big selling tuner amplifier is here now. The Module 80 has been tested and proven over the past six months on the continent with great success. This is why:—

- \* Superb selectivity
- \* Highly sensitive
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- \* Good stereo channel separation and definition

## Goodmans Module 80

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a new concept in low frequency measurement

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TIC Meters measure the inter-pulse period,  $t$ , of the input frequency, compute its reciprocal, and display the result as frequency to a high accuracy. Basic Sampling time is  $t + 1$  seconds. In the higher ranges,  $t$  is averaged over 10 or 100 periods.

#### Sampling Times

- 0.1 Hz—11 seconds
- 0.5 Hz—3 seconds
- 5 Hz—1.2 seconds
- 50 Hz—1.2 seconds
- 500 Hz—1.2 seconds

#### Features

- Up to 5000 : 1 improvement in Data Rate
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#### Applications

- Slow speed Tachometry
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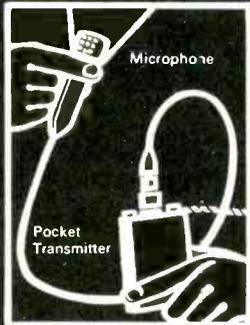
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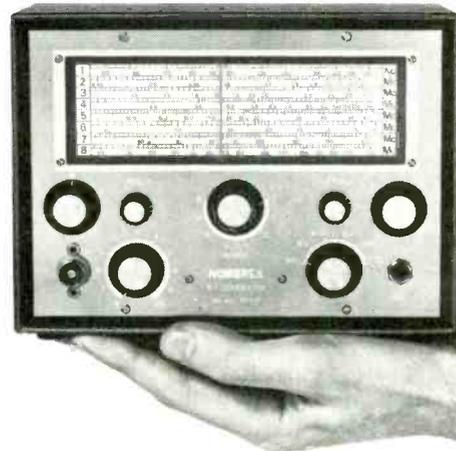


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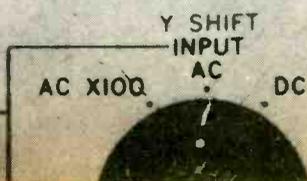
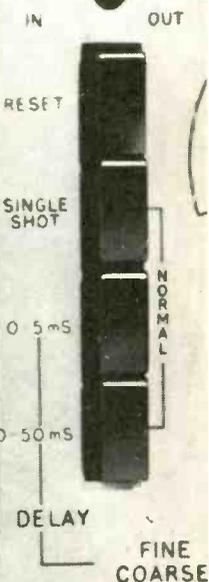


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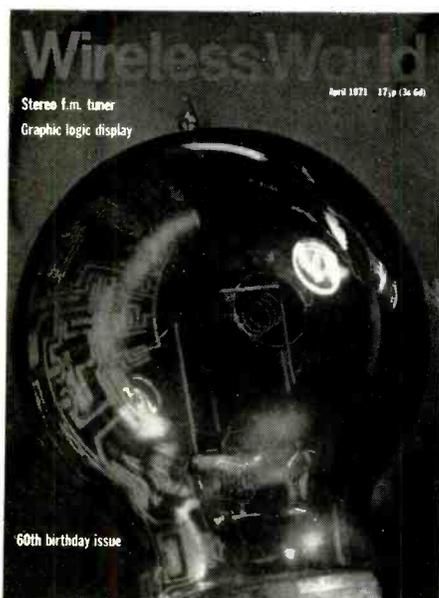
# Wireless World

Electronics, Television, Radio, Audio

Sixty-first year of publication

April 1971

Volume 76 Number 1426



The reflected image of an integrated circuit on the gettering of an R valve on **this month's cover** symbolizes the advances in technology during *Wireless World's* 60 years. (Photograph by Paul Brierley)

## IN OUR NEXT ISSUE

**Artificial vision.** A microelectronic implant for directly stimulating the brain of blind people in order to restore some degree of vision; r.f. signals are conveyed by inductive-loop transmitters and receivers. Full description of Medical Research Council work.

Circuitry for a **five-channel stereo mixer** will be described along with constructional hints. The various high-quality amplifiers can be used as components for an audio pre-amplifier.

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Brief extracts or comments are allowed provided acknowledgement to the journal is given.

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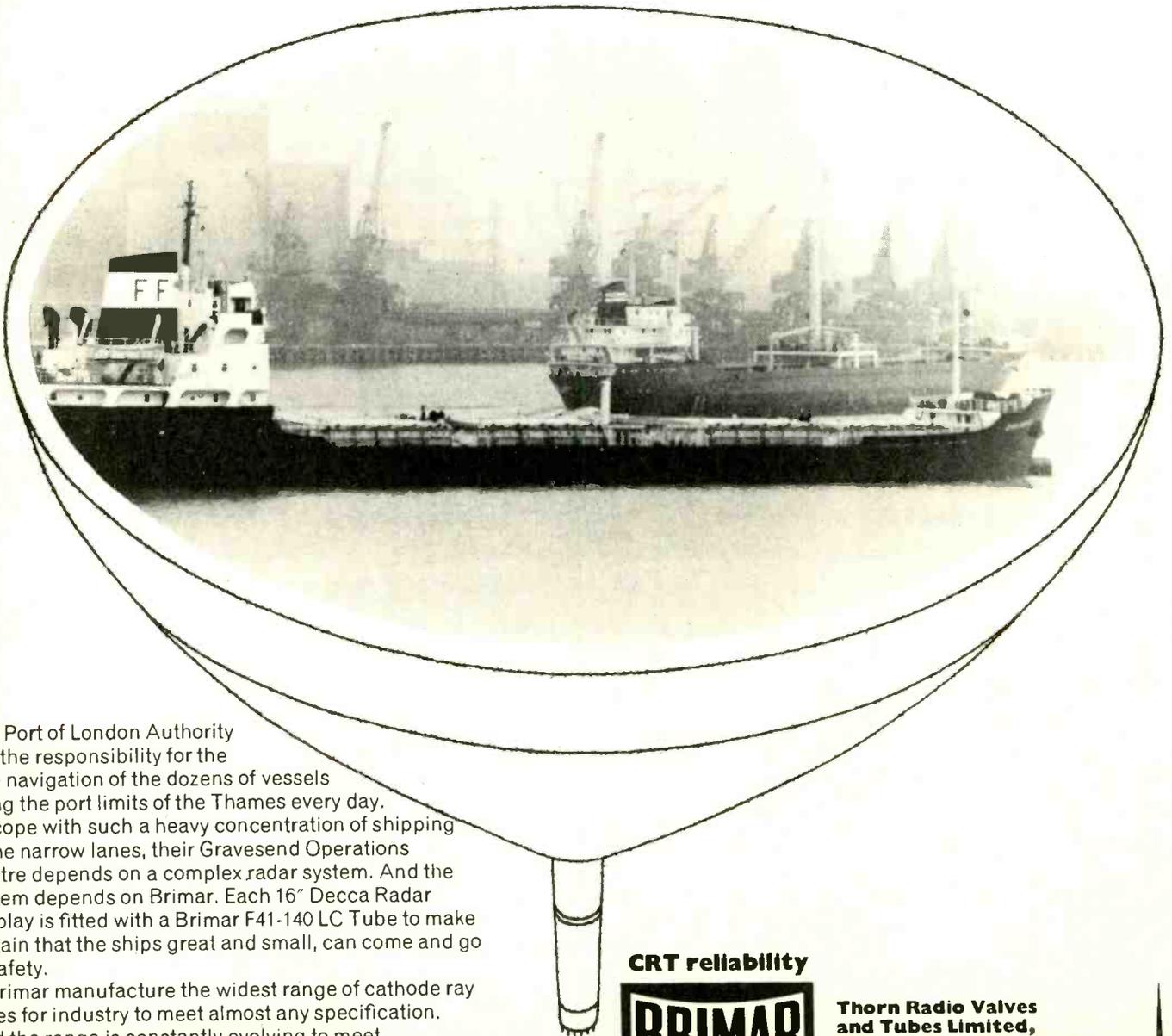
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# Wireless World

## Sixty Years

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*The editorial regularly presents us with an opportunity to express an opinion on some topic of interest. We rarely discuss ourselves and it is not our practice to have "guest editorials", but on this occasion, as we celebrate our 60th birthday, we have broken with tradition and have invited Hugh S. Pocock, F.I.E.E., to occupy the editorial chair once again. To him, more than any other one person, must go the credit for the development of Wireless World both in its formative years and in its later growth. He was editor from 1920 to 1941, then managing editor, and successively director, managing director and chairman of our publishing company until his retirement in 1962.*

*We also include in this issue several contributions reviewing progress during the past 60 years in various fields—audio, receiver techniques, basic theory, communications and radio propagation.*

*Looking back through the volumes of Wireless World we are conscious of the debt we owe to our contributors (some whose names became household words) whose knowledge and ingenuity has enabled us to maintain the ideals set in the first issue of W.W.—"This then is our policy: to be of use and interest to our readers, and through them to be a factor for progress."*

*We are also deeply grateful for the support given us by our advertisers and last, but by no means least, for the loyalty of our readers. Now we temporarily vacate the chair!*

When the *Marconigraph* was published by the Marconi Company in 1911 the intention was to provide a means of giving wider publicity to the Marconi System than had been possible through Guglielmo Marconi's lectures to scientific bodies and references in the Press. The circulation, however, was mainly amongst Marconi engineers and marine operators with a small readership amongst those interested in the Marconi Company as an investment or speculation.

After two years of publication it was decided to broaden the scope of the journal and put it on sale on bookstalls, with the new name of *The Wireless World*, the idea being to remove the impression that it was merely a Marconi publicity publication. Under the new title it was to continue to favour the Marconi System but to be broader in its attitude towards other activities outside the company.

In making preparations for the launching of the first issue of *The Wireless World* the Marconi Company put out an advertisement for an editorial assistant preferably with some knowledge of wireless. This advertisement caught the eye of one who was to be closely identified with the journal's fortunes for the next fifty years. His qualifications were a fluent pen (at that time), the holder of an Experimental Wireless licence, and that he had absorbed almost everything published on the subject at that date, although he admitted to dodging the mathematical analysis of the spark in Fleming's "Wireless Telegraphy".

Being accepted for the job he found himself installed in Marconi House in the

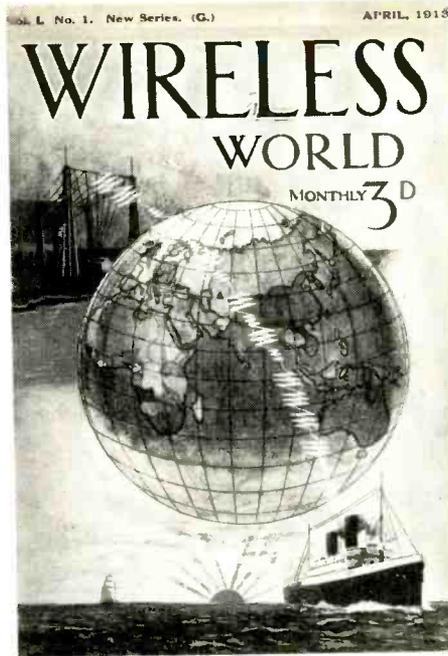
Strand, London. The site for the present Bush House, alongside, had just been cleared for building. Preparation for the first *The Wireless World* was being made but that was not a full-time occupation for the new recruit so to him was assigned, as an extra duty, the editorship of the new publication "The Year Book of Wireless Telegraphy and Telephony".

The Postmaster General had granted a good many Experimental Wireless licences by this time and Gamages produced a directory of them and ran a department to supply experimenters with the gear they required. The Wireless Society of London (which became the Radio Society of Great Britain) was a focal point and the then editor was keenly supported in the idea of fostering amateur interests. *The Wireless World* became the official organ of the society and published, in full, its lectures and discussions.

With the outbreak of the first world war there was strict censorship, experimental licences were withdrawn, and our publishing activities greatly circumscribed. We were, no doubt, the first wireless journal to have to submit material to censorship and we well remember taking our 'copy' to Whitehall, where the chief censor, F. E. Smith (later Lord Birkenhead), dealt with it personally.

A commission in the Royal Engineers with wireless and intelligence duties at home and then overseas meant a break in association with the journal until late in 1920, when the invitation, sent to Baghdad, to return to occupy the editorial chair of *The Wireless World* was a rewarding prospect.

When the present Editor recently did me the honour to invite me to make some contribution to the 60th anniversary number he said he had in mind that (for a very short time, I presume!) I should be back in the editorial chair and contribute a guest editorial.



The four-colour cover of April 1913 "Wireless World".

As that is the nature of the invitation it gives me every excuse to adopt the editorial 'we' as we proceed and we propose to confine ourselves mainly to touching on certain events and outside influences which have affected the journal's career.

With the lifting of censorship after the first world war a wealth of material became available for publication. The general availability of the valve provided great scope for inventors and experimenters.

Naturally, with such a promising field, *W.W.* did not long remain without competitors and a number of new journals appeared. The journal was taunted with its Marconi bias and consequent neglect of rival systems.

The Postmaster-General did not re-issue experimental licences despite the clamouring of the wireless societies and amateurs. Eventually the wireless societies decided to seek legal advice to obtain what clearly appeared to be their rights under the Wireless Telegraphy Act. Then something occurred which was to prove of great importance to our future. We were telephoned one evening by a press friend and told that a rival radio publisher had put out a news item to the press that £500 was being offered to the Wireless Society of London to assist in its legal show-down with the Post Office. A prompt telephone to the home of the manager of Marconi publications procured the authority to make a similar offer if we felt it necessary in the interests of the journal. So, the next morning our offer and that of the rival publication both appeared in the press. Nothing very interesting about that, one might say, but it had its repercussions! That morning we received a summons from the managing director of Marconi's, (F. G. Kelloway, a former Postmaster-General) to attend his office with the publisher. The Post Office had apparently

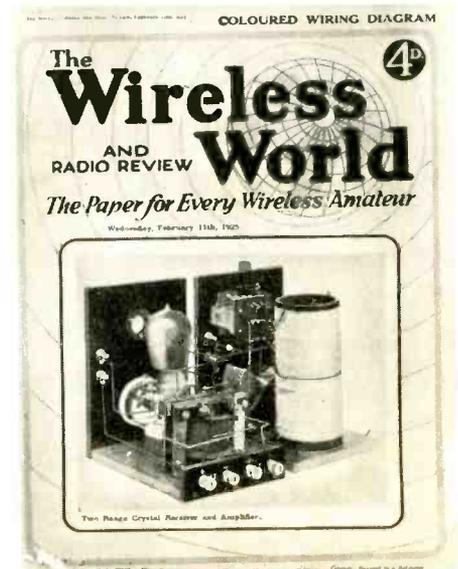
taken the line with the Marconi Company that it could not continue the present negotiations for wireless station contracts while the company's publication supported an attack on the Post Office monopoly. The outcome of that stormy interview was the decision to find a buyer for the offending *Wireless World*. This is how the journal came into the fold of Iliffe & Sons, Ltd in 1924. Though respecting our former proprietors, we welcomed the change wholeheartedly because it gave us editorial independence and we could no longer be charged with bias, while we had the very important advantage that we now had the resources of a top publishing house ready with financial support and experience of publishing. One of the first moves was to change the format to suit rotary presses for a much increased printing order. The competition from other journals in the field remained intense but we were able to hold our own and establish a reputation for sound designs for constructional articles and all round technical reliability.

We remember the occasion when the first issue under our new proprietors was on the machines we spotted a letter from a reader which expressed his disapproval of stunt circuits, but this appeared as STunt circuits. At that time our rival publishers gave a serial number to the constructional designs which they published and prefixed the number with ST, being the initials of the designer. With alacrity the printing machine was stopped and the offending capital letters reduced to lower case.

The most active competition in the field eventually closed down and those journals of the rival group which continued did so under other publishers. It is interesting to recall that we later received from the former proprietor of the Radio Press (John Scott-Taggart) a generous tribute to *The Wireless World*. (We hope he reads this in his Beaconsfield retreat!)



The cover of the first issue of the journal under its original title.



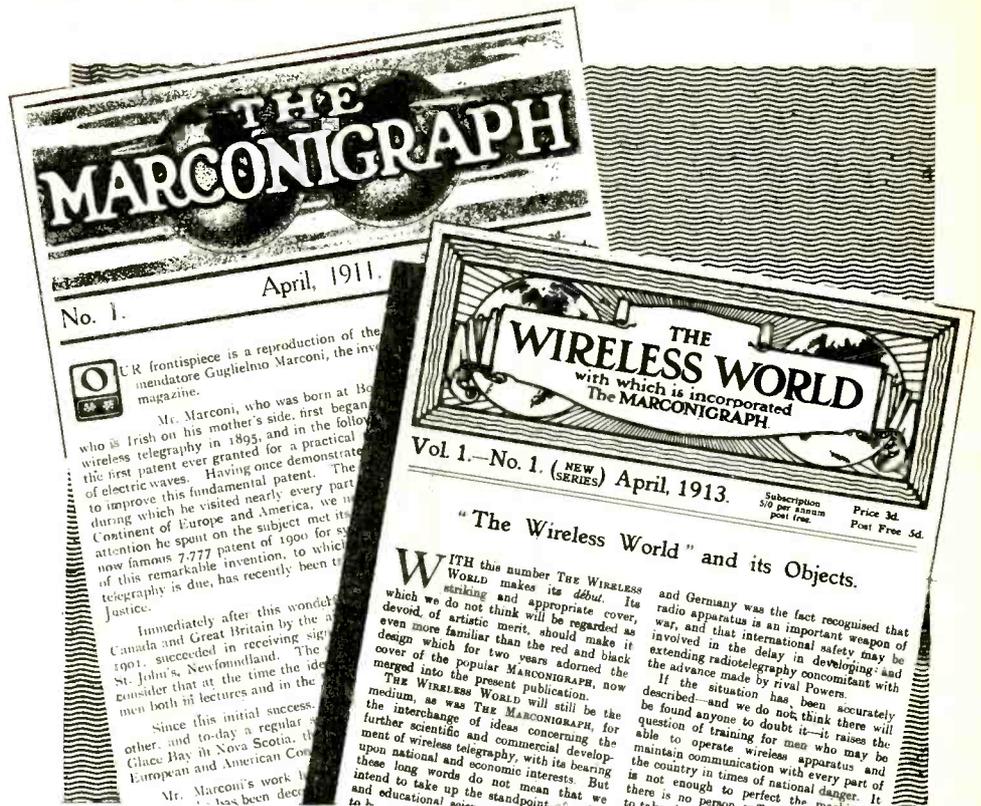
Constructional feature in this 1925 issue was a two-band crystal receiver and amplifier.

Hardly had the competition of the Radio Press faded out than we were confronted with another problem. The B.B.C., which had launched *Radio Times*, now produced a new journal *World Radio*. Profiting from the monopoly of their own programmes, they were able to obtain, by exchange, advance details of a wide selection of foreign programmes for publication in *World Radio*. This at a time when there was very great interest in receiving the foreign transmissions and designers of receivers here competed to achieve a degree of selectivity which made it possible to sort out the individual programmes. Next, *World Radio* added to its contents technical articles and constructional designs, carried a sub-title 'The Technical Journal of the B.B.C.', and competed with us for contributors. In addition to protesting, which seemed to have little effect, we felt we had to take steps to safeguard our position especially when the B.B.C.'s use of the microphone to publicize its journals was taken into account.

That is why *Wireless World* began the very expensive policy of breaking the monopoly by publishing foreign programmes, as well. Both journals became unprofitable and eventually *World Radio* discontinued publication of technical articles of the type we objected to and we agreed in return to discontinue foreign programmes. In a general agreement between the B.B.C. and the Press the B.B.C. undertook to confine its future publishing activities to what was 'pertinent to the service of broadcasting'. *World Radio* was closed down at a later date and in any case could hardly have continued to obtain advance foreign programmes as the cloud of war in Europe darkened.

As events seemed to be moving towards war we felt that there could soon be an urgent need for people skilled in the very field for which our journal catered. We believed our readership would be the ideal medium through which to recruit for such services. So we launched a 'Wireless World Register', inviting our readers, who would be ready to give their services in an emergency, to complete a form giving such particulars as we thought would be most useful. Having got the approval of the Services we published the form in the journal with the address side carrying O.H.M.S. (on the recommendation of the Admiralty). We believe that such a permission had never been given previously to any technical journal nor has it been granted since, as far as we are aware. There was a rewarding response from our readers and the completed forms went to the Wireless Telegraphy Board as a convenient clearing house. Perhaps this resulted in our getting less credit for our enterprise than would otherwise have been the case. The register proved very valuable especially in meeting the need for radar personnel.

The imminence of war made things very difficult for us. It is well known that technical journals depend very largely on their advertising pages for a healthy existence. Manufacturers were now so overwhelmed with orders for war needs that they saw no purpose in advertising and



Our object "to be of use and interest to our readers, and through them to be a factor for progress" remains unchanged.

we suffered badly. So unpromising was the position that a boardroom decision was made that we should close down. That might well have been the sad end of *Wireless World* but fortunately our board of directors was not composed of Medes and Persians and they were prepared to reverse a decision. We produced facts and figures to show how *Wireless World* could be expected to continue, even profitably, by changing from weekly to monthly publication with a corresponding reduction in paper and printing costs and a reduced staff which was already inevitable with departures of a number to the Services. Actually, with the change to monthly publication, we never looked back.

From this point onwards in our history the editorial 'we' should be taken to include successive editors, the late H. F. Smith, and also F. L. Devereux, two names which will always be associated with the very best that we have been able to put before our readers issue by issue through the years. We are proud to have been followed by men of such outstanding qualities. Throughout its history *Wireless World* has enjoyed the co-operation of a loyal and efficient editorial team and the journal's success must, of course, be credited to them and to our many outstanding contributors, both staff and outside, who have devoted their energies to the needs of our readership. Our present editor, H. W. Barnard, carries on the tradition with that dedication and competence which can be expected from one who has devoted the whole of his working life to *Wireless World*.

It would seem appropriate here to make reference to the transfer of Iliffe's to new proprietors. A good many years ago Iliffe's, being a private company, was

attached, for convenience, to a public company (the Amalgamated Press) then controlled by Lord Iliffe with his partners. Some time later Amalgamated Press was sold to I.P.C., the *Daily Mirror* Group, and we believe the new proprietors only later discovered how important an acquisition had come their way with the Iliffe journals. There was much reorganization and change, but *Wireless World* together with other electrical and electronic publications was gathered as one unit which still continues as a distinct entity constituted much as it was when we vacated the chairmanship of the unit to rest from our labours some eight years ago.

In "The Torrington Diaries"\* which record the travels through England of John Byng (later Viscount Torrington) the author states, in his introduction, "If my Journals should remain legible, or be perused at the end of 200 years, there will, even then, be little curious in them relative to travel, or the people: Because our Island is now so explored: Our roads, in general, are so fine; and our speed has reached the summit". This he wrote during a tour in Lincolnshire in June, 1791. But we can have sympathy with John Byng for how could we, at the time the *Wireless World* was launched, foresee the future through successive stages of the invention of the valve, with all its applications, short wave communication, radio telephony, broadcasting, radar, television, the transistor and the employment of electronic devices in almost every human activity. And we do not think anyone today would venture to suggest that we have "reached the summit"!

\* The Torrington Diaries (Eyre & Spottiswoode)

# Loud and Clear

## Developments in audio over 60 years

*remembered by F. L. Devereux, B.Sc.*

Of necessity the Editor has had to comb the park benches for someone long enough in the tooth to remember when *Wireless World* began and who at the same time was engaged in sound recording—albeit in the humble capacity of holder of the hot flat-iron near the wax cylinder while his father made records of piano playing.

Yes, home recording was well established before the first issue of *W.W.* made its appearance on the bookstalls, and my father's Edison-Bell phonograph boasted an exponential horn, sapphire styli for recording and playback and a hill-and-dale groove—recently revived in the Telefunken-Decca video disc. But wireless signalling was in a much more primitive state. There was no broadcasting as the term is now understood; indeed, wireless telephony, except for a few sporadic experiments of limited range and duration, was unknown.

When *Wireless World* began, signals were in morse—nice digital stuff. All you had to do to get the message clearly was to make it loud enough to stand above the threshold of background noise. Power at the transmitter and *sensitivity* at the receiver (amplification was to come later) were the first essentials in getting the signal from point to point; then it was a case of cutting down ambient noise in the receiving room until one could almost hear the blood circulating in the ear. My own dodge was to retire to an unventilated but heavily damped clothes closet, floor area 4ft X 4ft, with sufficient air for nearly half an hour, at the end of which time noisy breathing drowned all but signals from Eiffel Tower and Poldhu.

---

**F. L. Devereux** retired in 1965 after more than 40 years with *Wireless World*, including eight as Editor. In 1917, at the age of 17, he went to Parkeston Quay, Harwich, as a laboratory mechanic in the Board of Invention & Research engaged on anti-submarine methods. He later joined the Navy as a midshipman and after demobilization in 1919 went to Birmingham University where he graduated in physics. Before joining *Wireless World* in 1923 he spend a short time in industry.

In those days when the very idea of a loud-speaking telephone seemed like science fiction, the most sought-after piece of equipment was the Brown reed-driven headphone. This highly sensitive earpiece, with an aluminium cone diaphragm and a slack gold-beater's skin surround, was designed by S. G. Brown (honoured also as a pioneer of the gyro-compass). It had micrometer screw adjustment of the air gap between reed and electro-magnet and was a great advance on run-of-the-mill iron diaphragm types then current.

Having made the signal audible the next step was to try to amplify it so that several people could hear it simultaneously without diluting it further among several headphones. Brown was again to the fore, first with a truncated conical horn added to the reed movement and later with current amplification by attaching a button microphone to the reed.

It is significant that at this time any device for augmenting the signal after detection was termed a 'note magnifier'. Oh, happy days, when all one had to design for was a single frequency! Success was measured by the amount of noise the device produced, and if a few harmonics crept in so much the better—the note was "crisper" and probably easier to read. (Incidentally, I wonder how many present-day pilots and R/T operators see any incongruity in the expression 'How do you

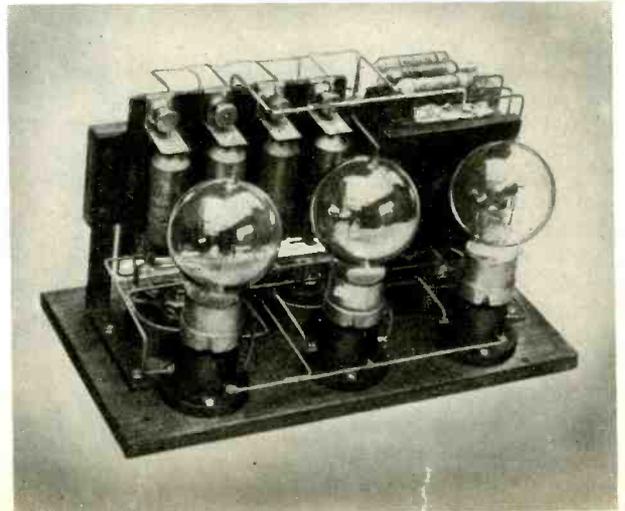
read me?' As a fugitive from the morse era, whenever I hear it I see in my mind's eye the written message and the pencil flying across the signal pad at 20 w.p.m.)

Many mechanical devices as well as electrical were developed to augment the sound, e.g. the Brown 'Frenophone' in which pressures from a reed unit were made to control the friction between a rotating glass disc and a pad to which the apex of a cone diaphragm was attached; the Johnsen and Rahbek system where the friction between a partially conducting rotating drum and a band brake was augmented by electrostatic forces at the interface; and the aptly named 'Stentorphone' in which compressed air was released through an electromagnetically controlled valve.

The signal was by now quite definitely LOUD, and to the excitement induced by the exercise of these new powers of amplification was added, in the early 1920s, the thrill of hearing voices and snatches of music through the babel of morse. And that's when our troubles really began: the primitive digital days were at an end and, headed by Fourier, analogue methods were rearing their ugly heads.

Meanwhile during the 1914-18 war, the thermionic valve (originally partially gas-filled) had got rid of its early flatulence and with the emergence of the hard-pumped 'R' valve took over entirely

Wireless World's first R-C coupled amplifier caused consternation among transformer manufacturers.



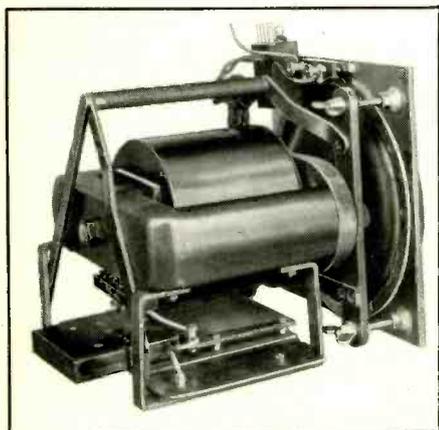


moving-iron variety—telephone diaphragms driving the fashionable swan-necked horns and reed mechanisms for the larger paper cone diaphragms. Both introduced prodigious asymmetrical distortion, proportional to the inverse square of the varying air gap. But who cared? The music kept good time, and by Edison-Bell phonograph standards the quality was quite acceptable.

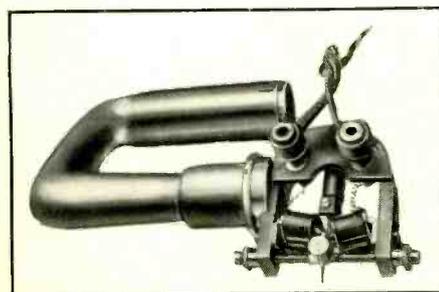
At this time and for years to come the B.B.C. set a standard of quality which was streets ahead of the capabilities of commercial receivers to reproduce. Protagonist in the drive for better sound quality was the B.B.C.'s first chief engineer, P. P. Eckersley, whose energy and wit did much to stir the listeners from their indifference and the industry from its lethargy. His campaign was powerfully reinforced by the introduction in 1924 of the Marconi-Round-Sykes moving-coil microphone in place of carbon-granule types.

Returning to loudspeakers, the Baldwin balanced armature unit and the Western Electric 'Kone' showed how non-linearity distortion could at least be reduced and bass response improved, and there was a final fling of the moving-iron principle in the so-called 'inductor dynamic' unit with motion parallel instead of normal to the magnet pole-pieces. But this was a last despairing effort to stem the advancing tide of the moving-coil, patented in 1874 by Siemens, again by Lodge in 1898, and brought up to date in 1924 by Rice and Kellogg of the GE Company of America.

No single step in the progress of sound reproduction has ever equalled that from



A new standard of performance was set in 1933 by the Voigt unit with a 20 kilogauss magnet.



A gramophone pickup of 1925. "Tons per square inch on a refined macadamised roadway".



Of 19 pickups tested by Wireless World in 1929 only these six (BTH, Brown, Burndept, Igranic, Magnum, Webster) recorded any output above 4kHz!

moving-iron to moving-coil, or fired so much enthusiasm. Those who were first privileged to hear the results felt compelled to spread the good tidings, and mass meetings were organized by the redoubtable Dr. N. W. McLachlan in London, by Dr. F. W. Lanchester in the Midlands and by *Wireless World's* Assistant Editor, F. H. Haynes at the leading radio societies. At last the good things which for years the B.B.C. had been wasting on the desert air could be appreciated and we entered the first golden decade of high-quality sound, culminating in the work of P. G. A. H. Voigt, whose many-sided genius showed that the lily could indeed be painted. Through the medium of his domestic corner horns with

their massive 20 kilogauss field magnets and twin diaphragms, and with a temerity which smacked of *lèse-majesté*, he disclosed faults in the B.B.C.'s own transmissions. I well remember turning up at his home in South London for a demonstration of his latest model, only to be told that it was 'off'. "They're still using that mic. with the shrieking 6-kilocycle frontal cavity resonance that I complained about last week". He was right, of course, and that microphone was subsequently taken out of service.

Thus is progress made, sometimes waiting for the man, sometimes for the means. The detailed design of stereophonic recording on disc, as it is used today, was worked out and patented in 1933 by A. D.

Blumlein, but at that time discs were made of heavily loaded shellac and, to quote Stuart Black (*W.W.* December 1943) "We get our music by scraping a steel point carrying some tons of weight per square inch over what is virtually a refined macadamized roadway". Not until the advent of vinyl co-polymers and microgroove recording could Blumlein's ideas be fully exploited.

And speaking of Blumlein reminds me of another story. When high-definition television started in England in 1936 all the high-quality enthusiasts were agog with excitement at the prospect of unlimited bandwidth on v.h.f. On medium waves the B.B.C. did not modulate above 10kHz because of international agreements on channel spacing. Sure enough, when the Alexandra Palace station opened up the improvement in quality of the accompanying sound channel was so marked that readers were demanding constructional details for v.h.f. receivers—just to demonstrate to their friends the virtues of unlimited bandwidth. At an I.E.E. discussion on the design of the A.P. transmitter several speakers rose to thank the B.B.C. for acceding to the public demand for greater bandwidth. To which Blumlein replied that he was sorry to disillusion the gentlemen concerned, but the improvement had nothing to do with bandwidth; for some time the B.B.C. had been in process of redesigning its microphone 'A' amplifiers and it just so happened that the new equipment, with greatly reduced non-linearity distortion, had been put into service for the first time at Alexandra Palace. And wasn't it P. P. Eckersley at another I.E.E. meeting who

said: "The wider you open the window the more dirt blows in?"

When one is stirring up memories of technological progress, why is it that people so often keep popping up among the hardware? W. S. Barrell of E.M.I.: "When I want to impress other people I play Tchaikovsky but when my staff ask my opinion of their latest improvement I always insist on Bartok". G. A. Briggs, who came into the loudspeaker business from the textile industry and who, if he brought a pair of cloth ears with him when he started, must have quickly used them up for diaphragm surrounds, for no one can so unerringly detect a false sound—and that without benefit of scientific aids (who is likely to forget his demonstration, at the start of the Festival Hall recitals, of the qualities of loudspeaker enclosures using nothing more sophisticated than a carpenter's mallet?). And C. E. Watts: I once made the mistake of saying to him that I did not see the sense of square-wave testing since such sounds were not to be found in nature, or for that matter in music. On my desk next morning I found a beautiful photomicrograph of three consecutive square-groove traces (triangular, of course, with constant velocity recording). Attached was a compliments slip endorsed "From Danse Macabre". I found out later that C.E.W. had spent most of the night chasing up and down the groove of that well-known test-piece of the period until he found what he sensed must be there.

Here I think we are getting near to the gist of the matter. When you know all about the physics of vibration and its transmission through the air—even about

the mechanism of the cochlea of the ear—you do not yet know the first thing about sound which is the *perception* of vibration. After all the trouble with Fourier and sine waves we are back where we started with the digital spikes of trains of nervous discharge travelling in times of the order of milliseconds along multiple paths, many of them redundant, to be subjected to correlation processes of the order of micro-seconds between the two halves of the brain—then probably to be overridden or ignored altogether by the recipient. As with sight the human capacity for instantaneous attention to detail is limited. Who knows or cares if there is a cut-off at 8 or 18kHz when the back desks of the violins are out of tune or the woodwind is dragging its feet?

Looking into the future it would seem that startlingly new technological advances will be few and far between, though it is always possible that some simple improvement awaits discovery under our noses. One need only cite the electrostatic loudspeaker which seemed to have been fully exploited by Hans Vogt in the late 1920s until Prof. F. V. Hunt and his colleagues at Harvard, after thorough mathematical analysis, showed thirty years later that what seemed to be inherent distortions could be removed by the simple expedient of working under 'constant charge' conditions.

Much remains to be done in the field of psychoacoustics, to find how judgments are conditioned by previous experience, why the mind accepts the false as the norm and often rejects improvement for no other reason than unfamiliarity. I am old enough to remember that switches marked 'mellow' had to be fitted to many broadcast receivers after the makers had tried to give the public better high-frequency response.

#### Envoi

As I drift, in the sixth age of man "into the lean and slippared pantaloons, with spectacles on nose" I know that if my old age pension does not run to a colour television licence I can still enjoy music, talks and plays—all those things that enter the mind's eye through the ear.

One small regret. The old *W.W.* Quality Amplifier was such a comfort on winter nights in my "den", but since changing over to integrated circuits I have had to buy an electric fire with *two* bars. We seem not so much to have miniaturized the watt as to have mislaid it altogether.



For his large scale lecture-demonstrations of high-quality sound on both sides of the Atlantic, G. A. Briggs might well have taken his motto from Danton: "De l'audace, et encore de l'audace et toujours de l'audace." Fears for their success were completely routea.

# Milestones in Receiver Evolution

**W. T. Cocking\***, an innovator in the field, recalls some of the highlights in radio and television receiver development

The real beginning of radio can be said to date from James Clark Maxwell's hypothesis, which he formulated in the latter half of the 19th century, that a displacement current (i.e., a changing electric field) could produce a magnetic field. He formulated this hypothesis in order to improve the symmetry of his equations relating electric and magnetic fields and he showed that, if it were true, one solution of them indicated electromagnetic waves travelling in space with finite velocity. There was at that time no way of proving or disproving his hypothesis, but later Heinrich Hertz succeeded in generating such electromagnetic waves. Still later, Marconi developed the elevated open aerial and the way was then clear for the practical development of wireless communication.

Sixty years ago when *Wireless World* started, all normal transmissions were by spark telegraphy using Morse code and receivers were very simple affairs using one or two tuned circuits with magnetic coherers or crystal detectors.

The triode valve had been invented

(1907) but few people had heard of it and it was not until World War I that it became widely known and was manufactured in any quantity. Practical radio-telephony had to wait for this moment because a source of continuous oscillations was needed as a carrier for speech signals and the valve proved the only suitable way of generating them.

The early '20s may be said to be the real beginning of radio as we know it today. There was military equipment available on the disposals market and valves could be bought. Receivers existed using up to five r.f. 'amplifiers' in cascade with special valves having contacts at each end of a glass tube for the filament and one on each side for the grid and anode (V24, Q or QX). Even with these low-capacitance types the stage gain was very low.

The normal receiving valve was the R type. This was a bright emitter with a filament taking about 0.6A at 5V. It had an a.c. resistance of about 40 k $\Omega$  with a  $\mu$  of about ten. All equipment was battery operated with a large accumulator for the l.t. supply and an h.t. supply of 60 to 120

W. T. Cocking's first contribution to *Wireless World* was in 1929, but it was not until 1936 that he joined the staff. During the war he served in the R.A.O.C. and R.E.M.E., attaining the rank of major, and from 1942 to 1945 was attached to the Ministry of Supply. After the war he was appointed editor of our sister journal *Wireless Engineer* and in 1965 became editor-in-chief of *Wireless World* and *Industrial Electronics* (successor to *W.E.*) which ceased publication in 1969.

volts from dry batteries. The usual valve receiver was a reacting detector with or without one transformer-coupled a.f. stage. Headphones were normal and many receivers had nothing but a tuned circuit and crystal detector.

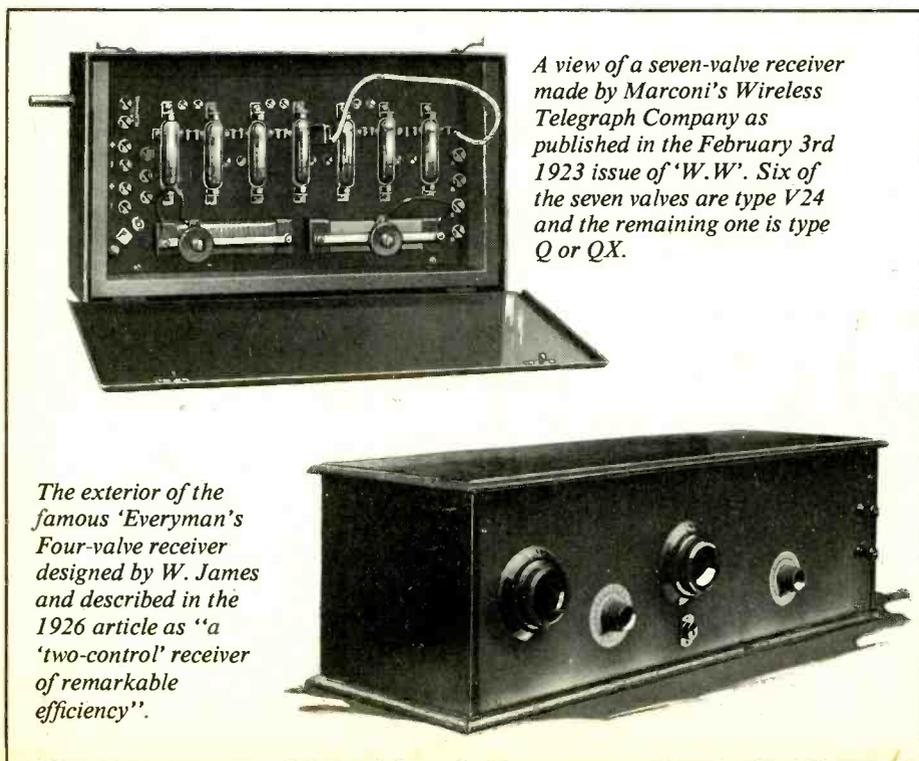
Broadcasting in this country began in 1922 and was the start of major development in receivers. Throughout the history of receiver development receiver designers have been quick to exploit, to the full, the potentialities of components available to them and the major advances have usually had to await component development—especially in valves.

The first valve improvement was the dull-emitter filament. The DER needed a 2-V supply only. Then came the '0.06' types, which took only 60mA at 3V and permitted the use of a dry battery for the filament supply; but this had too short a life to become really popular. The other characteristics of these valves were very similar to those of the R type.

Broadcast listeners began to dislike using headphones and the loudspeaker became popular. The results of the early loudspeakers driven by a grossly overloaded valve of the time were so horrible that power valves were developed. We should hardly call them so today, for few gave more than two or three times the output of an R valve! Then came the LS5 and LS5A. The latter, especially, did give useful power but needed 400V h.t. supply.

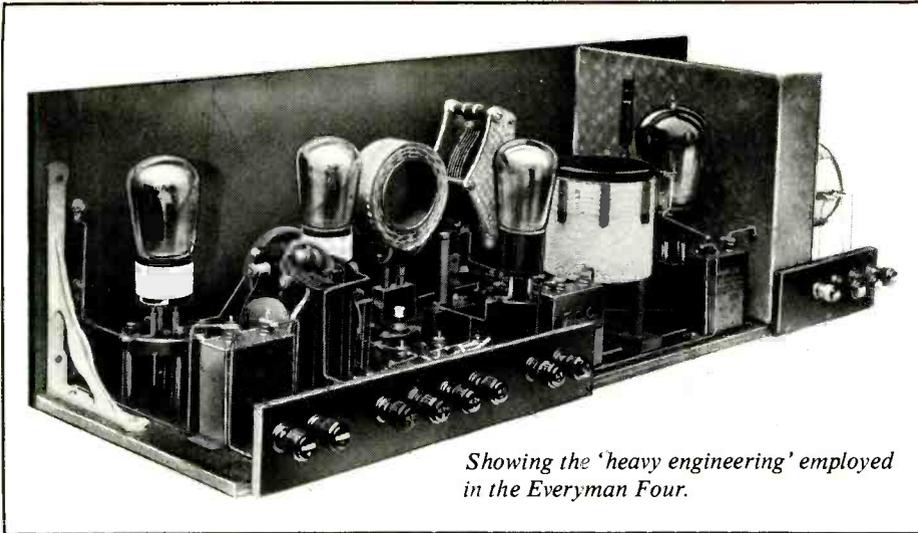
In this period there was great interest in the design of tuning coils. In the crystal-set days a solenoid of about 4 inches diameter and 12 inches long was used with a slider for tuning. Then variable capacitors (we called them condensers then) were used with plug-in coils of the honeycomb or Burndept types of windings.

*Wireless World* organized a competi-



*A view of a seven-valve receiver made by Marconi's Wireless Telegraph Company as published in the February 3rd 1923 issue of 'W.W.'. Six of the seven valves are type V24 and the remaining one is type Q or QX.*

*The exterior of the famous 'Everyman's Four-valve receiver designed by W. James and described in the 1926 article as "a 'two-control' receiver of remarkable efficiency".*



Showing the 'heavy engineering' employed in the Everyman Four.

tion among its readers who were invited to send sample coils. A prize of £5 was offered for the best. The results were published in the 17th, 24th Feb. and 3rd March 1926 issues (*W.W.* was then a weekly). At around the same period S. Butterworth published a series of articles in *Experimental Wireless & The Wireless Engineer* (April-July 1926) on 'Effective Resistance of Inductance Coils at Radio Frequency' and an abbreviated version appeared in *Wireless World* for 8th and 15th Dec. 1926 under the title 'Designing Low-Loss Coils'. Butterworth's work was of outstanding importance because for the first time it enabled unscreened air-core coils to be designed not merely for a required inductance but also for a required r.f. resistance.

At around this period, or a little earlier, the neutrodyne circuit was developed by Hazeltine in the U.S.A. and it enabled stable and useful r.f. amplification to be obtained. The typical American receiver of the period had two neutralized triode r.f. stages, triode detector and two trans-

former-coupled a.f. stages. There were three separate tuning capacitors, for ganging was still to come.

In *Wireless World* for 28th July and 4th August 1926 there appeared constructional details for a very famous receiver indeed, the Everyman Four. (Incidentally the original title was 'Everyman's Four-Valve', a prototype of which is kept at the Science Museum, London.) This had one r.f. stage neutralized with a DE5B valve ( $r_a=21 \text{ k}\Omega \mu=18$ ) with two tuned circuits. The coils were of 3-in diameter and  $3\frac{1}{2}$ -in long and had 74 turns of 27/42 Litz wire, each strand s.s.c. and overall d.s.c. The stable r.f. gain was 36.5 to 46 over the medium-wave band. The detector was an anode bend type using a Cosmos SP18 Blue Spot valve RC coupled to a DE5 output valve. The whole set took 10mA at 150 V and the output valve took 4.75 mA. The input power to the output stage was thus 710 mW and so the output to the

loudspeaker could hardly exceed 150 mW.

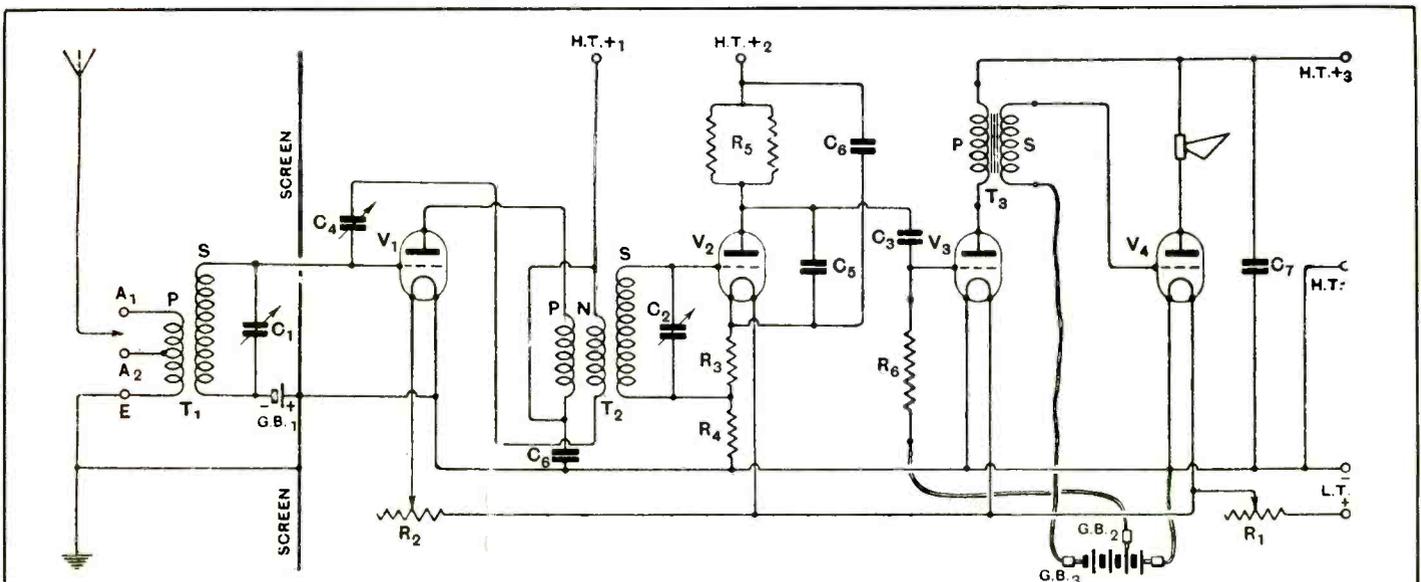
The performance of this receiver far outshone its competitors of the time and it set an entirely new standard of medium-wave broadcast reception. Its success depended largely upon the r.f. coil development referred to above, but also on the development of neutralizing.

In this country the neutrodyne never achieved the prominence that it did in the U.S.A., for it had barely reached here when the screened grid tetrode made its appearance and made neutralizing unnecessary. Shortly after this the output pentode appeared, and the introduction of the indirectly heated cathode made mains operation practicable.

From 1926 onwards valve development was rapid and receiver designs changed accordingly until in the early 30s the 'standard' receiver was mains operated with one r.f. stage, grid detector, and pentode output stage. Ganged tuning arrived. This was an obvious development, but its possibility depended on achieving much higher accuracy of manufacture of variable capacitors and it was not really satisfactory until this was done.

At this time the increasing number of broadcasting stations in Europe called for a great increase of selectivity in receivers. At the same time, much more attention was being paid to quality of reproduction. This led to the use of coupled-pairs of tuned circuits operating as bandpass filters ('Band-Pass Four' *W.W.*, June and July 1930) and then to the revival of the superheterodyne.

This first appeared during World War I, but was comparatively little used until the early '30s. The main reason for this was that it required more valves than the straight set and it was not until the advent of mains operation that many people could afford the extra power supply for these valves. Mains operation and better valves made the superheterodyne at last practicable but it was for a time held



Schematic diagram of connections. T<sub>1</sub>, aerial-grid transformer; T<sub>2</sub>, high-frequency valve transformer; T<sub>3</sub>, Ferranti 3.5:1 low-frequency transformer; C<sub>1</sub>, C<sub>2</sub>, 0.00027 mfd. square law tuning condenser; C<sub>3</sub>, 0.01 mfd.; C<sub>4</sub>, balancing condenser; C<sub>5</sub>, 0.0005 mfd.; C<sub>6</sub>, 1 mfd.; C<sub>7</sub>, 2 mfd.; R<sub>1</sub>, filament rheostat 2 ohms; R<sub>2</sub>, rheostat 30 ohms; R<sub>3</sub>, R<sub>4</sub>, fixed resistors 15 and 7.5 ohms; R<sub>5</sub>, 1 megohm each; R<sub>6</sub>, 3 megohms; GB, 15 volt grid battery.

A photograph of the original circuit, complete with caption, of the Everyman Four.

back by the problem of ganging the oscillator with the signal circuits.

An early design ('Super-Selective Six' *W.W.* June 1931) had separate controls for the two, but ganging was achieved quite soon ('Monodial', *W.W.*, April, 1932). Around this period, some unorthodox designs appeared. They achieved temporary prominence and then fell into disuse because of certain drawbacks, so they are really side shoots to the main line of development.

One such was the Stenode. This was a superheterodyne using a quartz crystal in the i.f. amplifier to obtain high selectivity together with tone correction in the a.f. amplifier for the severe sideband cutting. It aroused great controversy about the physical reality of sidebands and some of the claims made for it appeared to contravene accepted theory. The true explanation eventually appeared, but the Stenode never achieved any real popularity. One feature of it, the correction of sideband cutting by a suitable a.f. amplifier response, was adopted in the Monodial.

Another sideshoot was the Single-Span (1934). In this a high intermediate frequency (1.6MHz) was used with a fixed tuned input bandpass filter covering 150 kHz to 1.5 MHz. This enabled all tuning to be done by the oscillator ( $f > 1.6$  MHz) and eliminated ganging and waveband switching. Although satisfactory when first described it did not long remain so. The increasing numbers and powers of broadcasting stations soon produced so many whistles that it became impracticable without excessive refinement.

All this time spasmodic attention had been paid to quality of reproduction. An outstanding example was the Science Museum Receiver (*W.W.*, July and August 1930) which probably provided

the best reproduction of any equipment of its date.

An early Hi-Fi amplifier (this term had not then been invented) was the *Wireless World* 'Push-Pull Quality Amplifier' (May 1934). This had two PX4 triodes in push-pull driven with RC coupling from push-pull MHL4 triodes in turn driven by a concertina phase-splitter. The output transformer was specially designed for the job and contributed greatly to the performance. The output stage operated in a mode which might be called slight class AB. It was nearly class A but not quite. The amplifier gave 4 W output at an unspecified, but quite low, distortion level.

Some examples of it exist today and judged aurally the results compare well with much more modern designs.

In 1934 Black in the U.S.A. 'invented' negative feedback, primarily for amplifiers in cable circuits. This, as always, was a specialized field and it took sometime for the principle to work its way out for more general usage (*W.W.*, Nov. 1936), but within a few years it became common, although it was not at first always used to the best advantage.

At about this time the triode valve began to die as an audio output valve. Larger powers were being demanded and the biggest triodes (PX25, 25 W dissipation) were directly heated types. Indirectly heated cathodes did not seem to go with large power and in any case the pentode was more efficient (theoretical maximum 50% against 25% for a triode). The drawback of the pentode was that it introduced much more distortion than a triode and many quality enthusiasts would have nothing to do with it.

Here negative feedback came to the rescue and made pentode quality as good as triode quality and spelt the demise of the

triode output valve. It was eventually realized that the triode could be regarded as a pentode with 100% negative feedback from anode to screen! It was around this period (1936) that a.g.c., which is a form of negative feedback, came in.

To return to the early '30s, there were two other important developments. One was a kind of negative feedback, automatic gain control (*W.W.*, September 1932), although it was not then recognized as being such. It came into popular use long before true negative signal feedback. The second development (*W.W.*, Sept. 1932) was 'Ferrocart', which was the first iron-dust core for r.f. coils. Of German origin, it consisted of iron dust sprinkled on waxed paper which was then rolled and compressed to form a solid block which could be cut to the desired shape. Sometimes a ring core toroidally wound was used, at others the material was formed into E and I sections. It enabled a big reduction to be made in the size of r.f. coils and stimulated the development of the methods of construction and was soon replaced by cores of iron dust compressed with a binder into a solid. These were much nearer the cores of today.

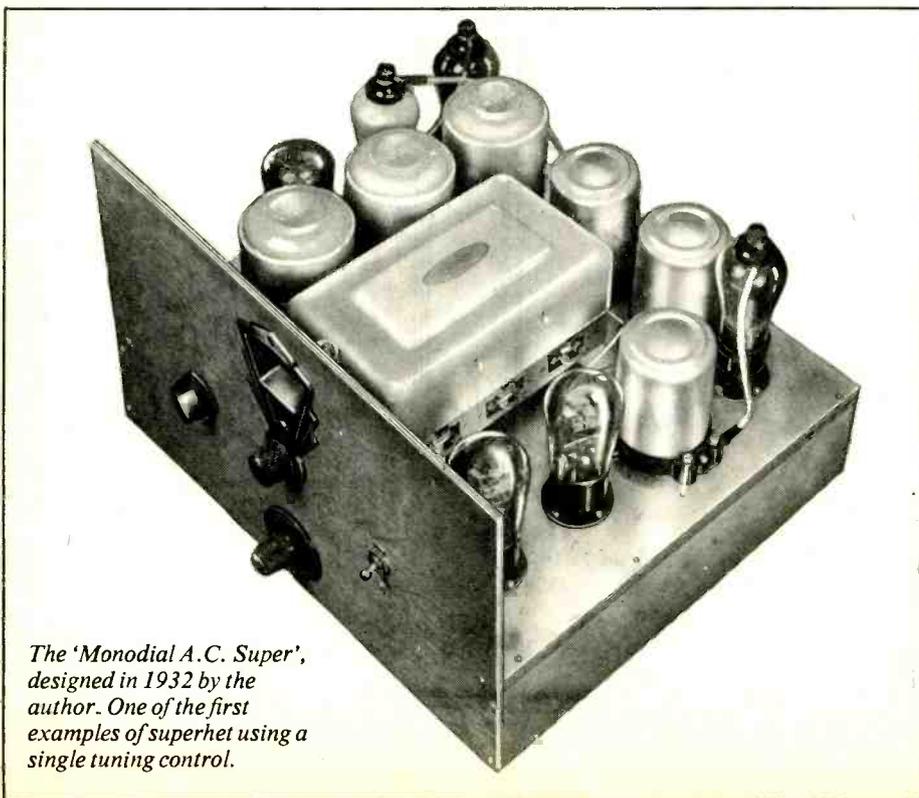
During the '30s there was intense activity in valve development not only in entirely new types, but in building multiple valves. The duo-diode-triode was one of the first, but by 1939 there were double triodes, double pentodes and triode-pentodes. The new types were mainly multi-grid ones for superheterodyne frequency changing. The first was the pentagrid or heptode and this was followed by the octode. A parallel line of development for the same purpose produced the triode-hexode, and triode-heptode.

Indirectly heated valves usually had 4V, 1A heaters, but in the U.S.A. 6.3-V, 0.3-A heaters soon became standard because of the introduction of car radio. American cars had 6-V batteries. When car radio came to this country a range of 13-V, 0.3-A valves was produced to suit our 12-V batteries. These could be series-connected for use in a.c./d.c. sets.

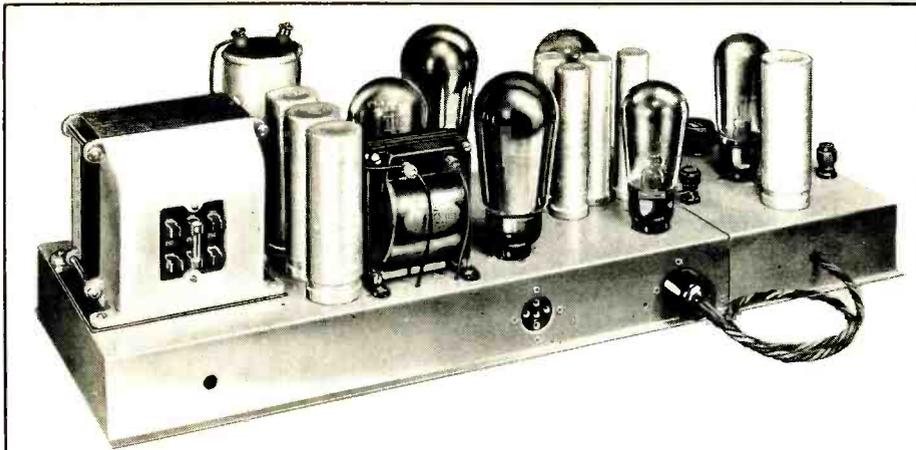
It was not long, however, before some degree of standardization with the U.S.A. occurred and 6.3-V, 0.3-A heaters became general and was followed just before the war with the all-glass construction of which the best known early member is the famous EF50.

The advent of television in 1936 shifted the emphasis in valve development to r.f. pentodes of high- $g_m$  and low capacitance. An early example was the TSP4 with  $g_m = 6\text{mA/V}$ , but this was quite soon replaced by the EF50 which had about the same mutual conductance but which was much better screened and was much smaller physically. Further valve development was interrupted by World War II and for a good many years the EF50 was the standard valve in radar receivers.

In the U.S.A. the small all-glass construction was adopted and later in the war such types were made here. These were more robust and had lower capacitances and higher mutual conductances, so that radar i.f. amplifiers, for instance, became smaller and more reliable. After the war



*The 'Monodial A.C. Super', designed in 1932 by the author. One of the first examples of superhet using a single tuning control.*



*'Push-pull Quality Amplifier' designed by the author in 1934. Some examples of this design are still giving satisfactory performance today!*

this form of construction continued and it is now employed in nearly all valves.

The next major development was the transistor in 1948. Of outstanding technical interest, it took a long time to have effect upon receiver design. The crystal detector, which is a semiconductor diode, was used in the very early days of radio. The commonest types were carborundum with a steel plate, zincite-bornite (Perikon) and galena with a catswhisker. During World War II it was revived as a centimetre-wave radar detector and properly designed with a pre-set capsule so that the user was presented with a little cartridge requiring no adjustment. Quite early on certain crystals were known to be able to generate oscillations under certain conditions, but little or nothing was known about how they worked.

As power rectifiers, copper-oxide and, later selenium, types were widely used from about 1930 onwards; little was known about how they worked, and their design was largely empirical. Nevertheless, they were very satisfactory.

The development of the transistor and the enormous amount of work on semi-conductors which preceded and followed it cleared up all these other matters. The first transistors were point-contact types and it was not until the junction transistor arrived and became manufactured in quantity that its effect became evident.

Apart from the computer, its first main application was to hearing aids. It had the outstanding advantages of much smaller physical size than a valve and of requiring no filament-heating power. It was also more efficient in that, because the minimum permissible voltage drop across the device could be under 1 V (with a valve it had to be 10 V or more) the h.t. supply power could be reduced.

In receivers, the transistor for the first time permitted the construction of a really practicable portable receiver. Portables were made and sold from quite early days, but they were large and heavy, had a poor performance and short battery life. The transistor changed all this as soon as types capable of amplifying and oscillating up to about 2 MHz became available. So much is this the case that the old

table-model receiver is now virtually obsolete. It is now the mains-operated radiogramophone or the battery-operated transistor portable. The development of ferrites, too, made the ferrite rod aerial possible and eliminated the cumbersome frame aerial from the portable. In the domestic market valves are now rarely used except in television receivers, but even here there are some which depend entirely upon semi-conductors.

During World War II, as we have said, valves began to get smaller. To match this components generally got smaller, too. The process of miniaturization had started. The transistor accelerated this and discrete components are now incredibly small by pre-war standards. But this is far from the end. Integrated circuits are with us and are coming into use in the domestic radio field. Even the variable-capacitor seems to be on the way out. It is starting to be replaced by a special semiconductor diode known variously as a varactor or varicap which has a capacitance dependent upon the voltage applied to it. At the moment, the only difficulty seems to be to manufacture diodes which all have the same capacitance at the same applied voltage; it is virtually the ganging problem again.

We cannot conclude without some mention of television. This started as a regular service in 1936 with one transmitter at Alexandra Palace, and after a short trial period the present 405-line system was adopted. Receiving cathode-ray tubes had a 12-in. diameter screen with electric focusing and deflection and operated at 4kV. By modern standards the tubes had poor focus and brightness.

Constructional details of a television set were given in *W.W.*, 2nd-30th July 1937. The receiver was of the t.r.f. type with three r.f. stages, diode detector and one video stage. The power supplies formed a large part of the cost and bulk, for three separate ones each with its own mains transformer were needed. Supplies of 250V for the receiver proper, 1000V for the time bases and 4kV for the c.r.t. Commercial practice of the time was similar, but a few manufacturers employed the superheterodyne. A major difficulty

was to obtain stable and high r.f. gain with the necessary bandwidth, because the valves available were not really suitable.

In the next two years, great improvements were made, partly because better valves became available and partly because of a change to magnetic deflection and focusing of the c.r. tube. To reduce costs the 9in. tube was adopted, and at least one set (Murphy) sold for £30. A second constructional receiver was described in *W.W.*, 29th June-20th July 1939, which took great advantage of these developments and gave a performance greatly superior to that of the first.

The war interrupted television, of course, and it was not until 1946 that transmissions started again. The post-war receivers naturally followed the immediate pre-war practice and the whole trend was to 9in. tubes with magnetic focusing and deflection. The service restarted on 7th June 1946.

Another constructional receiver was described in *W.W.*, Jan.-Dec. 1947. This was probably unique in that it included full constructional details of deflector and focusing coils, the reason being that such parts were not available on the retail market at that time. The receiver was initially of the t.r.f. type, but later a superheterodyne of much higher gain was described as an alternative. The e.h.t. supply, which was still no more than 5kV, was obtained from the line flyback using a voltage-doubler with selenium rectifiers.

War-time, and early post-war, valve developments made a big difference to television receiver design, especially on the r.f. side. The development of ferrites, too, had a big effect, for it so greatly reduced the losses in line-scan transformers and deflector coils that it permitted a further development—the energy-recovery scanning systems. These are now universal, but on them depended the practicability of wide-deflection angles and, hence, large screen tubes and the higher voltages needed for adequate brightness with them. The period 1947-1957 was an exceptionally interesting one in development.

Then, of course, came a 625-line system on v.h.f. and, finally, colour and a constructional receiver (*W.W.*, June 1968-June 1969), again appeared.

In this article, some may feel that undue stress has been laid upon designs for the home constructor. There is a sound reason for quoting these, however, which is that much more information about them is available than of commercially produced receivers of the time, especially in the early days. The heyday of the constructor was in the '20s and early '30s. After that, it became less popular as receivers became more complex, but the commercial pattern changed also and it gradually became more expensive to make a receiver than to buy one!

The demand for constructional articles fell off but the old saying, "An ounce of practice is worth a ton of theory", is still true. It is not that theory is unnecessary. It is more necessary than ever. It needs the practice, however, to drive it home and make one realize to the full what it means.

# Radio Wave Propagation

## Ten more years

by R. L. Smith-Rose, C.B.E., D.Sc., F.C.G.I., F.I.E.E.

In the 50th birthday issue of *Wireless World* a review was presented on the development of our knowledge of the manner in which electromagnetic waves travel over the earth's surface and through the lower and upper atmospheres, and of the experience which has resulted in the development of practical communications on a world-wide basis<sup>1</sup>. It is the purpose of this article to review the progress that has been made during the past ten years, taking note, as appropriate, of the associated developments in radio astronomy and space communications.

Because electromagnetic waves travel, subject to conditions of absorption, refraction and reflection, not only round the surface of the earth but also into the surrounding space, it has long been recognized that international collaboration is essential if confusion and interference are to be avoided in the practical development of communications, navigational guidance and the satisfactory broadcasting of sound and television programmes. It is on account of the international aspects and the need to avoid a chaotic state of radio interference, that organizations such as the International Radio Consultative Committee (C.C.I.R.) and the International Union of Radio Science (U.R.S.I.)<sup>2</sup> are continuously in operation to guide and control both the practical development and the scientific research associated with this subject. The introduction of sound and television broadcasting, advanced radio aids to both aerial and marine navigation and, more recently, the pursuit of research in radio astronomy and the space around us, have all served to emphasize the need for such international co-operation.

### Influence of terrain on wave propagation

The development of medium-wave broadcasting has, for many years past, stimulated the continued study of the effect of the electrical conductivity and dielectric constant of the ground on the propagation of radio waves over the earth's surface. The C.C.I.R. has produced, and published from time to time, sets of curves showing the decrease of field strength with distance from the transmitter which is assumed to be radiating a power of one kilowatt. It is further assumed that both the sending and

receiving stations are on the ground and that the waves travel over a smooth homogeneous earth, neglecting any effect of the troposphere. Five sets of such curves were revised in 1970<sup>3</sup> and have recently been published from Geneva. These sets of curves relate to four different values of the conductivity of the earth over which the waves travel, while the fifth set is appropriate to the much higher conductivity and dielectric constant of sea-water. Individual curves relate to a series of frequencies between 10 kHz and 10MHz.

Recommendations covering the use of the curves emphasize that they should be used to determine field strengths only when it is known that ionospheric reflections at the frequency under consideration will be negligible in amplitude. An example of such application is given as propagation in daylight at frequencies between 150 kHz and 2 MHz, and for distances less than about 2000 km. These sets of curves continue to form the basis of international discussions on the siting of broadcasting and other radio transmitting stations.

Methods have also been developed for computing the propagation conditions over ground paths of mixed electrical constants, such as are encountered in travelling from dry sand to a fresh-water lake, or from normally moist soil to sea-water of greatly increased conductivity. Similarly the effect of a variable terrain, including hills and mountain ridges which may be regarded as sharp irregularities in relation to the wavelength, has been studied in considerable detail to obtain information on such effects required for the planning

of broadcasting and other services which depend on ground-wave transmissions. Renewed emphasis on the desirability for further study of this subject has arisen during the past decade by the need for such earth-bound services to share some of the bands of frequencies with space telecommunication systems. Practical experimental work in this field has been conducted in parallel with a large amount of theoretical study, so that the combined results may be used in planning radio systems and predicting their performance with a good measure of reliability.

### Ionospheric research and long-distance propagation

Apart from national internal services, the major portion of the world's communications is conducted in high frequency radio waves, taking advantage of the appropriate reflection of such waves from the several regions of the ionosphere. It is now over 45 years since the classical experiments of Sir Edward Appleton demonstrated the existence of ionized regions in the earth's upper atmosphere, which reflect radio waves within suitable bands of frequencies, resulting in the transmission of the waves all round the earth's surface.

Continuous research carried out in various countries has shown that the frequencies of waves that can be so reflected depends upon the density of ionization in the atmosphere at heights from about 100 to 400 km above the earth's surface. It was established some years ago that this ionization process is dependent upon the intensity of emission of ultra-violet radiation from the sun. Furthermore, physicists have known for a long time that this emission is subject to variation on a basis with a period of the order of 11 years. As a result the range of frequencies or wavelengths which can be used for the world's long-distance services is much greater during a period of maximum solar activity than in the corresponding period about 5 years later.

In a previous article<sup>4</sup>, mention was made of the co-operative scientific study which was conducted on a world-wide basis of conditions in the ionosphere during the period of maximum solar activity (1957-58)—the International Geophysical Year (IGY), as it was termed. With the

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**Dr. R. L. Smith-Rose**, who is 76, retired from the Scientific Civil Service in 1960 after 41 years' service. A graduate of Imperial College, University of London, Dr. Smith-Rose was superintendent of the Radio Division of the National Physical Laboratory from 1939 until 1948 when he became the first director of radio research in the Department of Scientific and Industrial Research (now the Science Research Council). He is a past president of the International Scientific Radio Union and is at present chairman of the Frequency Advisory Committee of the Ministry of Posts and Telecommunications, and secretary-general of the Inter-Union Commission on Frequency Allocations for Radio Astronomy and Space Science.

knowledge provided by astronomers that the mean period of the sun's activity is about 11 years, a similar and enhanced programme of studies of the ionosphere was planned and carried out during the period 1964-65, which was designated the International Quiet Sun Year (IQSY). During this period, for the first time observations at observatories on the earth's surface were supplemented by direct measurements of conditions in the ionosphere, made first with the aid of rockets and later by the launching of complete radio sounding equipments through and above the ionosphere.

### The topside ionospheric sounder

Prior to 1960 rockets and artificial earth satellites were already in use for the measurement of solar radiation and the study of its effect on conditions in the ionosphere.

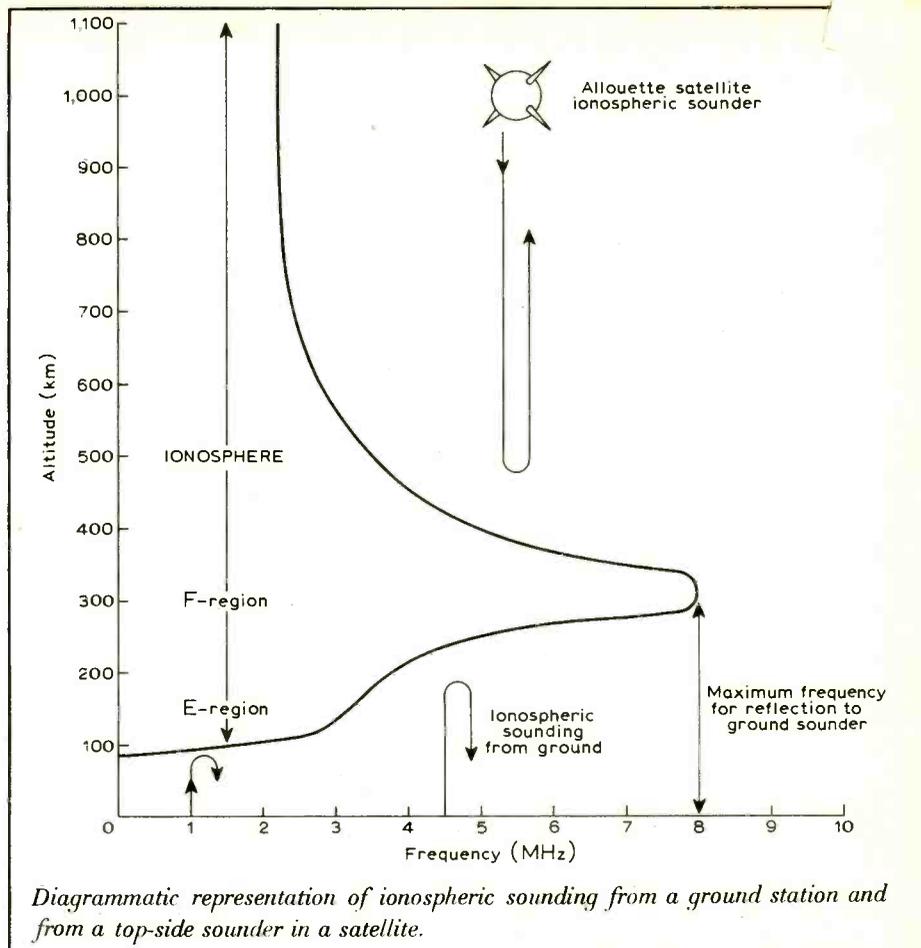
It was in September 1962 that a complete radio ionospheric transmitting and receiving equipment, known as the Alouette I top-side sounder, was launched into an approximately circular orbit at a height of about 1000 km. The frequency of the transmitter swept over the range 1 to 11.5 MHz in a period of eleven seconds, during which time the satellite had moved about 120 km; so that one complete ionogram was produced for approximately every degree of latitude.

The results obtained from this investigation have proved a most valuable supplement to the information provided by the world network of ionospheric sounding stations on the earth's surface. The fact that it was examining the properties of the ionosphere from above, in a virtually continuous world-wide orbit, brought to light some new and interesting points concerning anomalous geomagnetic conditions at the equator and the interaction between the previously identified radiation belts and the ionosphere below them.

The Alouette I satellite has provided a valuable series of observational emissions over a period of several years. It was still operating when, in November 1965, another artificial satellite—called Alouette II and also built in Canada—was launched into an elliptical orbit with major and minor axes of about 2980 and 500 kilometres respectively. This second satellite operates on command for about six hours per day, carrying out five experiments, which include a topside sounder, a radio-noise experiment over the frequency range 0.2 to 14 MHz, and the measurement of very low frequencies over the range 50 to 30,000 hertz. Both these and other satellites launched more recently have proved very successful in materially adding to our knowledge of radio transmission conditions at heights well above the ionosphere.

### Radio meteorology and the troposphere

For many years past the meteorologist has used radio sounding technique to give him detailed information of the temperature,



pressure and humidity changes in the earth's atmosphere up to heights of 10 km or more. In return, this information has proved invaluable in the planning and operation of radio communication services operating at very short—metre and centimetre—wavelengths. Under what is termed a normal or standard gradient of atmospheric temperature with height, such waves may travel in a path curved toward the earth at a radius of about four-thirds that of the earth itself. Variation of atmospheric conditions along the path may, however, change this to a greater or less curvature, including what is virtually rectilinear propagation<sup>5</sup>.

The development of direction finding and, later, radar techniques, has also enabled the radio scientist to explore wind movements up to the maximum heights in the troposphere, varying from 10 to 15 km depending on latitude and season. By international collaboration, a considerable amount of useful empirical knowledge has been gained from such combined radio and meteorological investigations in different parts of the world. But the search for a simple method of applying a knowledge of meteorological conditions to the determination of radio propagation has not led to very satisfactory results. In spite of such difficulties, however, the combined experience of scientists and engineers has enabled a certain amount of guidance to be made available to those responsible for the installation and operation of radio services at decimetre and centimetre wavelengths. A useful recommendation recently brought up to date by the

C.C.I.R. incorporates a revised set of curves relating field strength to distance of transmission for the v.h.f. (30-250 MHz) and u.h.f. (450-1000 MHz) bands. These curves display a statistical average of received field strength for 50% of the terminal locations and for periods of from 1 to 50% of the operating time. Associated reports enable the effect of changing the receiving aerial height to be estimated, and describe a method for determining the corresponding field strengths when the path of transmission is of a mixed land and sea nature.

But it is not only for the design and operation of radio systems with earth-bound terminals that a detailed knowledge of the effects of the troposphere is necessary. Modern developments of unmanned satellites in orbit round the earth for radio relay communication purposes also require a detailed knowledge of the propagation of radio waves through the non-ionized regions of the atmosphere, taking account of spatial variations of refractive index which can cause both refraction and scattering of the waves. With this type of work is associated an investigation of the absorption of radio waves by oxygen and water vapour of the variable densities encountered in the earth's atmosphere, and of the corresponding scattering of the waves particularly caused by various types of rainfall.

The development of telecommunications on an international scale depends to a major extent on an agreement as to the

type of investigations to be carried out and, particularly, on the nomenclature used in organizing the work and describing the results achieved. It has been evident for many years past that the propagation of radio waves of frequencies greater than 30 MHz is greatly influenced by meteorological conditions in the troposphere. In recognition of this the C.C.I.R. has drawn up a recommended list of terms used in the study of radio propagation through the troposphere. This vocabulary<sup>3</sup> was started nearly twenty years ago, and it has been constantly extended and revised as necessary at successive meetings of the international committee dealing with radio communications. Associated with the vocabulary are the agreed definitions of a basic reference atmosphere and the recommended formula for the radio refractive index. All these activities have done much to extend the successful application of the upper portions of the electromagnetic spectrum to practical use.

### Radio astronomy

It may not be out of place to conclude this review with a brief reference to radio astronomy, a science which has made

great advances during the period under review in many parts of the world. Excluding the relatively small activity in the field of radar astronomy, which uses a combined transmit-receive technique, the activities of the radio astronomer are confined to studying the natural radiations from sources in space, not only within the solar system but out to the limits of the explorable universe.

By the aid of either an extensive fixed aerial array or, more usually, of a large steerable aerial system, the astronomer is able to record and investigate the radiations emitted over the entire radio spectrum. So valuable has this work become in the past decade that a special international commission was set up in 1960, to review the requirements of the radio astronomer and to take all appropriate steps to ensure that his observations of these natural phenomena should be protected from interference by other services operating within the terrestrial environment.

In some cases the radio astronomer has identified specific emissions from natural phenomena, such as the radiation from neutral hydrogen in the frequency band 1400-1427 MHz. But more generally the

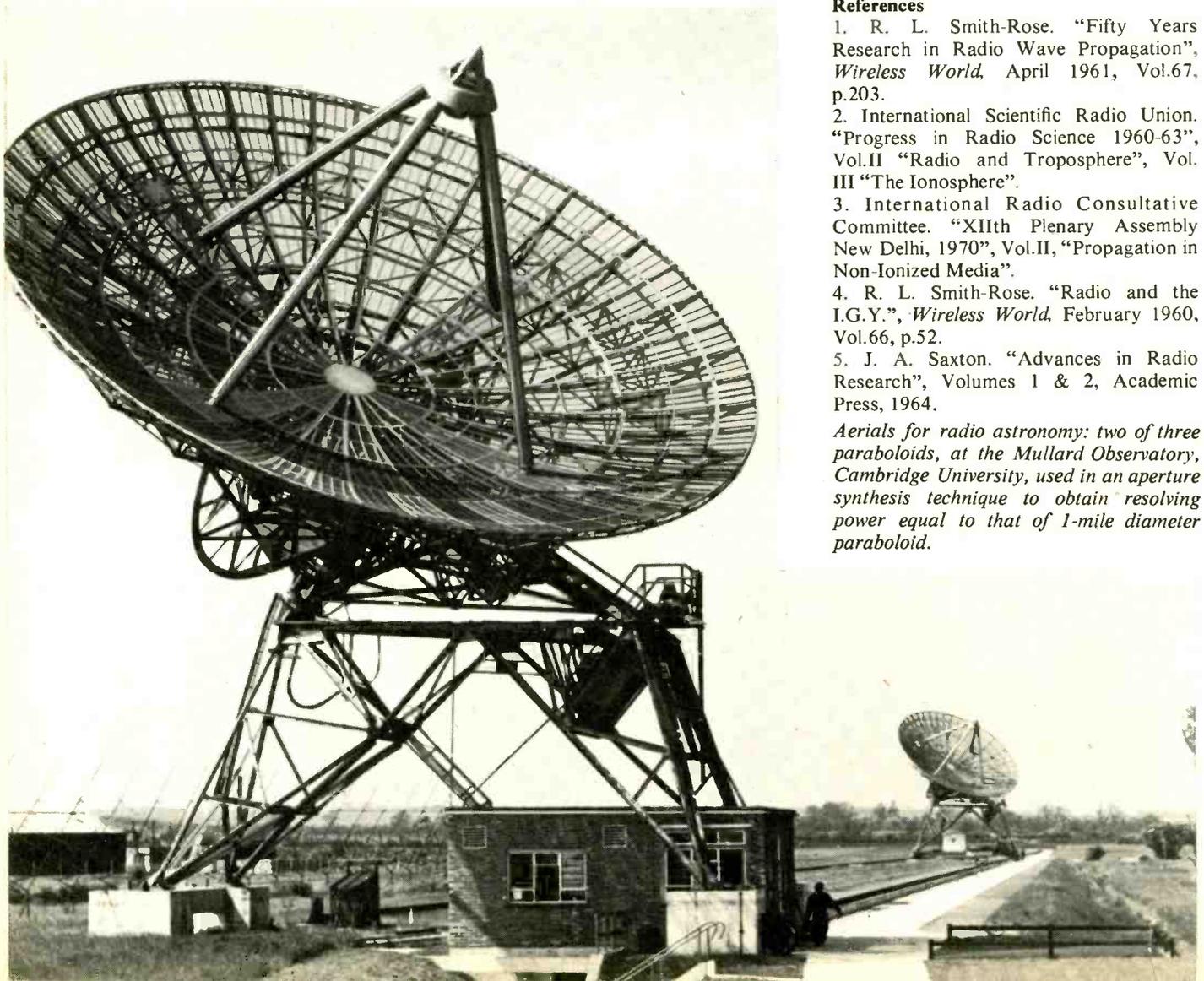
astronomers have sought protection from interference in a series of frequency bands at approximately octave intervals throughout the spectrum, so that they may conduct a co-ordinated long-term series of continuous observations of the phenomena which give rise to these radiations. The results so far obtained have materially added to our knowledge of the history of the universe which was already available from the much older work of the optical astronomer. While the major additions to our knowledge have been obtained from installations on the earth's surface, space radio astronomy has developed rapidly in recent years, culminating in the launching of the first Radio Astronomy Explorer Satellite for the observation of solar and galactic radiation free from the absorption caused by the earth's atmosphere.

It is perhaps of interest to note, in conclusion, that while up to a few years ago radio communications and control systems operated within the limitations of the earth's circumference, the modern astronaut seeks and receives a corresponding service which has already operated successfully at a range of about a quarter of a million miles—the mean distance between the moon and the earth.

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*Aerials for radio astronomy: two of three paraboloids, at the Mullard Observatory, Cambridge University, used in an aperture synthesis technique to obtain resolving power equal to that of 1-mile diameter paraboloid.*





indicate conclusively that the use of long waves in conjunction with high powers was the correct formula for long-distance communication. Thus by 1918 we find the Marconi 'timed spark' station at Caernarvon transmitting on 14,000 metres with a power of 200 kW, using a directional aerial which could radiate towards New Brunswick or the Antipodes. To this was added, in 1920, a long-wave 100 kW valve telephony transmitter which also established contact with Australia.

Ever since 1910 Guglielmo Marconi had had the dream of providing the British Empire with a chain of wireless stations to link its units together in a manner which would not be nearly so vulnerable as the cable circuits in time of war. Various governments, for various reasons, had procrastinated over this and the issue had not been settled by 1924, but matters then looked more hopeful and in fact, orders for long-wave high-power (1000 kW) stations had already been received by the company from Australia and South Africa. At long last, Marconi's cherished ambition was coming true.

Into this situation Marconi and his assistant C. S. Franklin themselves inserted what was very like a spanner in the works. Various experiments on wavelengths between 10 and 100 metres had been undertaken since 1917; these were primarily for short-haul links, but it had been noted that on occasion the signals, while dying out at a comparatively short distance were reappearing hundreds of miles away. Wireless amateurs too, having had these 'useless' wavelengths forced upon them, were reporting trans-oceanic ranges which occurred at some times and not at others. Marconi and Franklin, in 1923, conducted exhaustive experiments between a specially built short-wave station at Poldhu and Marconi's yacht *Elettra* and these fully confirmed the skip-distance effect. Further tests, using various wavelengths from 32 to 92 metres established a rough rule-of-thumb as to the best wavelength to use at a given time of day to reach a given destination; it was also established that Australia could be contacted with a fraction of the power used by the longwave giants.

This, then, was the nature of the spanner. Orders were on hand for two huge, expensive long-wave stations. But,

secreted in the company files, were details of an entirely new concept; the use of short waves which, by reason of the manageable dimensions of the aerial arrays, could be beamed to destination instead of being scattered broadcast. Should the long-wave orders be executed notwithstanding, or should the customers be informed of the new development?

The solution was not so simple as all that. Freak propagation conditions might account for the extraordinary ranges; time alone could tell whether this was so or not. Again, the short-wave beam equipment used had been strictly experimental; it all needed engineering. Was it justifiable to put forward a largely untried experimental rig against a proven (but very much more expensive) system?

Marconi had everything to gain by proceeding with the original order and then developing the beam system at leisure, producing it some years later. Characteristically, he did it the hard way and kept faith with his customers, the Australian and South African governments. He told them the exact situation, offered beam stations in lieu of the long-wave giants and let them decide for themselves. Both opted for the beam system, with Canada following suit. The British Government and the Post Office agreed to the provision of an Empire beam system, provided that the first circuit (between Canada and Britain) fulfilled the stringent conditions laid down. It did so, and by the end of 1927 the Empire beam service was in full operation.

Not the least remarkable part of the story is the technical feat of C. S. Franklin and his small team. With the decision made, Franklin had to engineer the experimental transmitter into production form, to modify valve design, to design the various aerial systems to operate on the wavelengths selected and to devise a means of conveying the output power of the transmitter to the aerial system without undue loss. All this was done in a matter of weeks—the last-mentioned by Franklin's invention of the concentric feeder, forerunner of the coaxial cable.

### Thermionics

Not all momentous steps are immediately recognized as such. An instance occurred

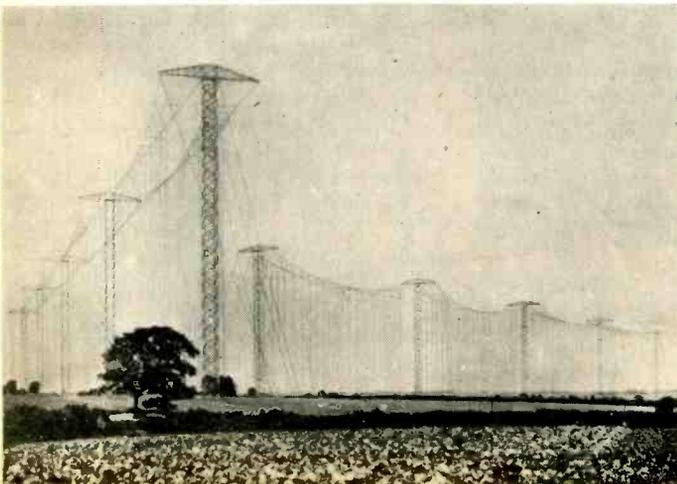
in 1904 when Dr. J. A. Fleming, Marconi's Scientific Adviser, utilized the Edison effect (originally noted 22 years earlier) to develop the first thermionic valve, the diode. Although not the first electronic device to be used (the 'Italian Navy' or 'Solari' detector of 1901 was a form of semiconductor rectifier) the Fleming diode was, nevertheless, the foundation stone of electronics as we know it today.

In 1906 Dr. Lee de Forest, of the U.S.A., set the cat among the pigeons by patenting a three-electrode valve, for which powers of amplification were claimed. Fleming was never a man to suffer rivals graciously and the Marconi Company promptly filed an action to restrain de Forest from manufacturing and to invalidate his patent, claiming then the diode constituted the master patent and that the third electrode, the 'grid-iron', was an appendage. The first lawsuit, heard in the U.S.A., gave Marconi's the verdict, but this ruling was overruled by another court. There followed an interminable series of legal actions which dragged on for years. Although World War I brought an easement of the situation, the wrangle was not finally resolved until the 1920s, when a compromise was effected.

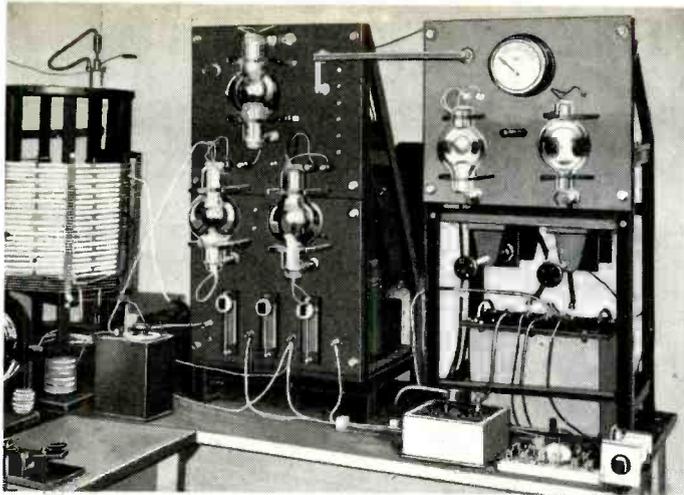
The circumstance was ironic in that, at the onset, the squabble was over a virtually useless device. The early triodes had practically no amplification factor and were, at best, temperamental performers. Their theory of operation was imperfectly understood and it was widely accepted that a 'soft' vacuum or a gas filling was essential. Not until 1911-12 when research by Dr. Irving Langmuir and others produced the 'hard' or high-vacuum valve, was the device transmuted into a really effective component. The importance of this work cannot be over-emphasized because, for the first time, a batch of valves with reasonably similar characteristics could be predicted and manufactured.

In 1913 came another tremendous technical advance when A. Meissner (Germany) patented the first thermionic generator. In this he was closely followed by Franklin and Round (Britain) and, a little later, by Armstrong and de Forest (U.S.A.) This discovery not only made wireless telephony a practical proposition (although it had been done before by non-thermionic means) but it formed a critical point in the history of electronics. Until that time there had been one broad highway and one only—wireless telegraphic communication. The development of the triode valve in its dual roles of amplifier and generator (a process accelerated by the Great War) brought a post-war diversification which has continued to this day.

Sound and television broadcasting, the present gramophone industry, the talking picture industry, public address systems, electronic navigational aids, television, radar and electronic test instrumentation are just a few of the roads which branched from the wireless telegraphy highway as a result of the



*Receiving station at Bridgwater, Somerset, typical of the Marconi-Franklin beam stations set up in 1926 for the Post Office.*



*2MT, the Writtle, Chelmsford, transmitter operated by P. P. Eckersley in 1921/2.*



*Lauritz Melchior broadcasting from the Chelmsford works of Marconi in 1920—two years before the formation of the B.B.C.*

development of the triode, and which are now arterial roads in their own right.

### Broadcasting

All these diversifications, like Topsy, 'just grew'. Sound broadcasting for instance, came into being after World War I largely by accident when engineers at Westinghouse in the U.S.A. and Marconi's in this country, becoming bored with reciting into their respective microphones for range tests, used gramophone records as interludes during which they could restore their vocal chords. To their surprise they found they had a small but coviferous ready-made audience of amateurs clamouring for more. In Britain, this situation led, via the Melba and Melchior concerts, the joyous 'send-up' approach of P. P. Eckersley and his team at 2MT Writtle and the sobriety of the original 2LO at Marconi House in the Strand, to the formation of an association of British radio manufacturers known as the British Broadcasting Company in 1922. (It became a Corporation in December 1926.)

Television had a rockier road to tread. Historically its concept (as a closed circuit system with wires as the transmission medium) pre-dates sound broadcasting by more than half a century. Its practical realization, however, had to await the development of the valve amplifier. For

example, the apparatus designed by Nipkow in 1884 did not come to fruition until 1926 when John Logie Baird, using the Nipkow system of spinning-disc scanning, with the indispensable additions of an improved photo-cell and valve amplifiers, became the first man to give a public demonstration of television pictures which had movement and a degree of light and shade in them. He also implemented a suggestion made by A. Sinding-Larsen in 1911, namely that the signals generated by the televising apparatus might be used to modulate a radio-frequency carrier wave.

Not all steps which are taken are forward ones. In 1907 Professor Rosing (Russia) had pioneered the use of the cathode-ray tube as a means of picture display and achieved still pictures of simple geometric shapes. In the following year A. A. Campbell Swinton (a Scot) outlined a proposal for an all-electronic scanning system; he followed this three years later (1911) with a more detailed account and set down the essentials of the modern camera tube. He also visualized the receiver as using a cathode-ray tube type of display. Nothing significant was done to develop such a scheme, however, until V. Zworykin (U.S.A.) patented his iconoscope in 1923 and P. Farnsworth, another American, was known to be working on his image dissector. Zworykin's camera tube embodied the

important inter-scan storage principle, which Farnsworth's did not. Neither inventor, it seems, knew anything about Campbell Swinton's proposal at that time.

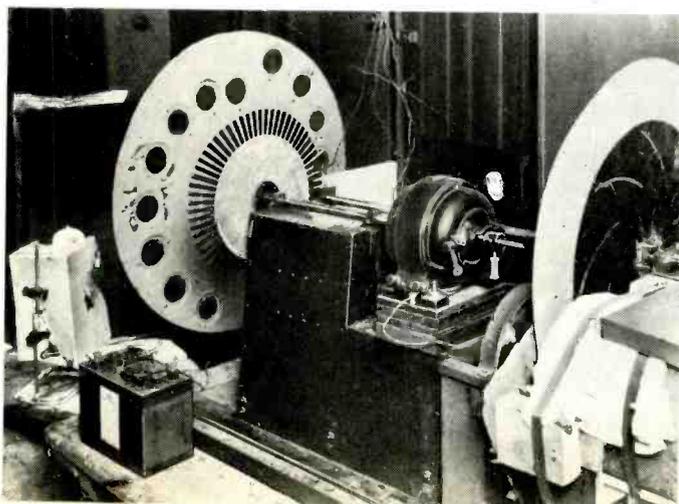
Both devices presented immense difficulties in manufacture and it was not for some years that a practical demonstration of either could be given. In the meantime Baird demonstrated, in rapid succession, the production of television pictures in darkness, colour television, and stereoscopic pictures and in 1930 began a limited experimental 30-line public service, using two B.B.C. medium-wave stations, one for vision, one for sound.

The publicity he gained in the late 1920s encouraged many of the bigger radio manufacturers to investigate the possibilities of television. Unfortunately almost all their effort was concentrated on various mechanical systems and in consequence a good deal of money went down the drain.

Electric and Musical Industries Ltd was one company which thoroughly investigated mechanical scanning, the EMI team being under the brilliant leadership of Isaac Shoenberg. By 1932 Shoenberg saw clearly that electronic scanning was the system for the future and intensive research was done on camera tube design, resulting in the Emitron tube. The team also developed interlaced scanning.

In 1934 the television interests of EMI were merged with those of the Marconi Company to form a new organization, the Marconi-E.M.I. Television Company; by so doing the skills of E.M.I. in video work were allied to Marconi expertise in wideband modulation and amplification (it had been realized for some time that the complex high-definition video signals would have to be transmitted at v.h.f.)

Two years later (1936) the Government-appointed Selsdon Committee recommended a public trial of the Marconi-E.M.I. system against a new high-definition system developed by the Baird Company. Transmissions on both systems, on a turn-and-turn-about basis, were radiated by the B.B.C. from Alexandra Palace, North London, for



*Using the Nipkow system of spinning-disc scanning, Baird gave public demonstrations of television in 1926.*

\*The partnership was dissolved in 1948.

several months from November 2nd, 1936. The Marconi-E.M.I. Company transmissions employed all-electronic scanning at 405 lines per frame, interlaced, while Baird used 240 lines sequential scanning; high-speed Nipkow discs performed the scanning process for televising individual subjects in the Baird studio, while an intermediate film system took care of larger scenes. By February 1937 the battle was over, the Marconi-E.M.I. system emerging the winner. Transmissions were, however, suspended from the outbreak of war in 1939 until June 7th, 1946.

### Electronics diversification

Over the years, from the mid-1930s to the present time, electronics research effort has mounted steadily; gone are the days of the lone-wolf investigator, his place taken by the mass-attack technique or by a sizeable team. The resultant multiplicity of inventions in a great diversity of directions makes it extremely difficult to select those which constitute the most significant advances from a list that would fill a book.

Which to mention in a limited space? The development of v.h.f. and microwave techniques immediately clamours for attention. So also does the study of the ionosphere and troposphere; the development of the super-het and frequency modulation by Armstrong; the pioneering of radio astronomy by Jansky and others; these and dozens more cry out for comment. But, in the space available, three innovations, radar, the transistor and satellite communications stand out as vital steps in the technology.

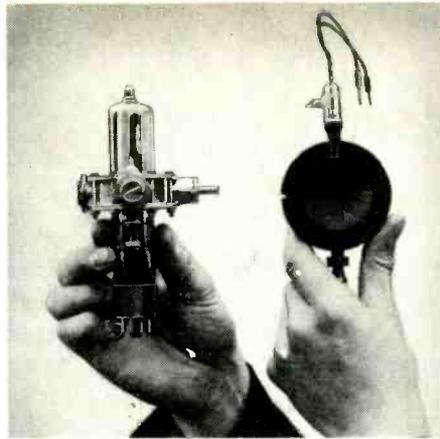
### Radar

Hertz in his earliest experiments had shown that wireless waves could be reflected from a metallic screen; Tesla, in 1900, suggested that such waves might be used to detect moving objects, while in 1904 Hülsmeyer actually patented a rudimentary form of radar. Then the dark ages set in; not until 1916, when Marconi and Franklin were working on two metres, was interest revived in reflected waves. This was referred to by Marconi in 1922 in the course of a speech to the American Institute of Engineers when he gave a remarkably accurate prophecy of how the detection of remote objects might be accomplished.

The practical development of what is today known as radar began with the work of Appleton, Barnett, Briet, Tuve and others in ionospheric sounding; the two last-mentioned seem to have been the first to have used pulsed transmissions for the purpose.

The first proposal for the use of pulsed transmissions for radar purposes was made in 1931 by Butement and Pollard of the Signals Establishment at Woolwich. They built a 50 cm equipment using a rotating beam and succeeded in detecting echoes from objects at a range of about 100 yards, but neither the War Office nor the Admiralty were interested and so the work was abandoned for lack of support.

Robert Watson-Watt's paper "Detection and Location of Aircraft by Radio



*Examples of early klystron and magnetron tubes.*

Methods", produced in 1935, is well known, as also is his work on the development of a practical radar system, for which he was subsequently knighted. Watson-Watt's work in this connection provided the chain of radar stations of various types established in Britain by the beginning of World War II. In 1939, the development of the resonant cavity magnetron by Randall and Boot was another momentous forward stride, in that it made high-power centimetric radar a reality. Centimetric Air-to-Surface Vessel radar, when introduced, brought a dramatic increase in U-boat destruction. Airborne radar for the interception of night raiders and a form of secondary radar which identified 'friendly' from 'hostile' aircraft were also introduced early in the war.

In the post-war years, to date, both ground and airborne radars have come into wide usage on civil aircraft and have become an indispensable air traffic control aid at busy airports.

### The transistor

Beyond question no other single component developed in the post-war period has influenced radio communications (and electronics in general) to anywhere near the same extent as the transistor. The rectification properties of certain crystals were known in the early 1900s and in 1906 Dunwoodie introduced the carborundum detector. Various crystal rectifiers have been used over the years, reaching their hey-day in the sound broadcasting boom of the mid-1920s but surviving in enormously improved form to the present day.

In 1911 Dr. Eccles announced the oscillating crystal but experimental interest in this lapsed for many years until *W.W.* in 1924-5 published various articles dealing with the phenomenon.

Contrary to general belief, it would seem that the first solid-state amplifiers as we know them were not, after all, invented at Bell Telephone Laboratories. The distinction goes to Dr. J. L. Lilienfeld who filed a patent entitled "Method and Apparatus for Controlling Electric Currents" in Canada in 1925. Although his description of how his invention

worked is wrong, the drawings show it to be a form of n-p-n transistor. Just why the potentials of this device were not explored is not known, but the next announcement of a solid-state amplifying device is that of Shockley, Bardeen and Brattain of Bell Telephone who, in 1948, announced their development of the point-contact transistor, for which, subsequently, a Nobel Prize was awarded.

### Communication via satellite

The evolution of radio communications has, until fairly recently, been largely a matter of coming to terms with the ionosphere. No single frequency band is ideal for all purposes; each has its advantages and limitations. The l.f. and m.f. bands are, for example, suitable for long- and medium-range transmission, but can accommodate only one information channel per carrier. H.F. transmission systems also provide long-range facilities but can carry only a few channels. The very high frequencies and ranks above give progressively greater channel-bearing capacities but the direct signals are limited in range to line-of-sight and therefore, as a generalization, could only achieve long-distance working by employing a series of installations as point-to-point links†.

In 1945, Arthur Clarke, in a *W.W.* article entitled "Extra-Terrestrial Relays" prophesied that future progress lay in the direction of artificial earth satellites equipped with receivers and transmitters. This forecast came to practical fruition in the early 1960s and, in particular, in 1962, when the Telstar satellite was the medium of a successful series of communication transmissions (including television programmes) between the U.S.A. and England; subsequent launchings of synchronous satellites have provided a 24-hour service to the point where communication via satellite is commonplace. By such means, v.h.f. and the higher frequencies have been freed from their former short-range limitations and can have their multi-channel capabilities exploited for long-distance communications. To date, however, h.f. still remains the backbone of long-distance radio communications, on economic grounds.

### Knowledge versus wisdom

As is, I suppose, proper in a technical journal, the discussion has been confined to technical advances. Whether the application of such discoveries has made for a better world is not the business of the engineer and the physicist. Or is it? How far are we responsible for the part radio communication has played—and continues to play—in the destruction of lives in war-time? Again, sound and television broadcasting has affected the pattern of living to a profound degree, the extent of which we cannot fully appreciate. Its vast potential for moulding social behaviour and world opinion is, however, all too often concentrated on holding a distorting mirror to the faces of its audience. Are our hands clean in this respect?

† Scatter links are exceptions.

# Basic Theory Since 1911

## Some of the controversies that have raged over the years

by "Cathode Ray"

One basic theory which, I suspect, is held by many today is that way back in 1911—and no doubt for many years after—'wireless' was developed in scientific darkness on a 'try it and see' basis, quite different from the modern approach. A look through the early issues of *The Wireless World* would surprise and enlighten any who think that the theoretical basis of those far-off days was good only for a giggle.

The first thing that pulled me up when I did some browsing was an article by H. M. Dowsett in the May 1913 issue, 'Molecular Structure of Insulators'. That title wouldn't look at all out of place in 1971. It was quite a simple treatment which one could hardly improve upon for present-day beginners. After all, though Fermi-Dirac statistics have entered our ken in the meantime and would have to be included in any full treatise on insulators, one doesn't need them in an elementary picture.

Prof. G. W. O. Howe's treatises on electromagnetic waves (Nov. 1913) and aerial capacitance (1915) were *not* elementary. I suspect the Editor would quickly turn them down today as too mathematical. All levels of intellect were catered for in *The Wireless World* and besides theory articles for beginners a free wireless instruction course was offered by a certain Robert Baden-Powell. (*The Wireless World* itself was almost free, by our standards, being 5s a year—12 issues—including postage.)

One notes with interest that as early as 1911 someone was asking 'Does Wireless Affect the Climate?' Nuclear explosions, moon rockets and the Concorde have now largely taken the place of wireless as a scapegoat for unusual weather. It is worth noting that 60 years ago Campbell Swinton delivered a paper which included a detailed outline of essentially the present system of television, using cathode-ray cameras and receiving tubes.

Among people still living, John Scott-Taggart seems to have been the first to appear in *The Wireless World* (Dec. 1914), followed closely by H. S. Pocock (Feb. 1915). P. G. A. H. Voigt, better known now for audio, wrote on reflex receivers in Dec. 1921. It was not until 23rd Aug. 1923 that I came in, but still far

enough back to make it hard for me to imagine what it is like to start now. Is there anything to compare with the excitement we had on first hearing spark morse signals, very memorable still after 50 years? Observation of the way children now accept colour television leads me to doubt it. But such reminiscing is an intolerable self-indulgence by the aged, and anyway is outside my present brief.

As we look back have we any ground at all for a feeling of superiority about our present-day theory? Well, however sound people like G. W. O. Howe may have been (and he long continued to get us on the right tracks with his famous editorials in *Wireless Engineer*, the sister journal of *W.W.*) one must admit that many in the amateur fraternity (and some even of the professionals) were a bit hazy. The principles of tuning caused a lot of difficulty. One of the main problems at the relatively low radio frequencies used was interference due to atmospheric, or Xs as they were called. All sorts of ideas for tuning them out were thought up, usually in ignorance of the fact that interference of this kind, being almost aperiodic, shock-excited the tuning circuits at their own natural frequency.

'Cathode Ray' started his inimitable series of expository articles in *Wireless World* in 1934. Many of these articles have since been published in book form—'Second Thoughts on Radio Theory' and *Essays in Electronics*. For many years the identity of 'Cathode Ray' was closely guarded but eventually it became known that it was M. G. Scroggie who has contributed to the *Journal* under his own name since 1923. When accepting our invitation to contribute to this 60th birthday issue he wrote 'Your letter started me on to calculating how much I have had published in *W.W.*; the result is 725 contributions (including reviews and letters over 47 years, totalling 1,400,000 words. Can anyone top this record?' If we exclude members of the staff, the answer is no! Mr. Scroggie graduated at Edinburgh University and was chief engineer of Burndept Wireless Ltd from 1928 to 1931 and since then, except for the war years, has been a consulting radio engineer. During the war he was first in charge of the early CH radar station at Pevensy, Sussex, and then at No. 9 R.A.F. Radio School before going into the Air Ministry.

The urge of amateurs to achieve the maximum ranges of communication was often unaccompanied by a corresponding grasp of the factors involved, and a belief grew up that results were essentially outside the scope of theory and were maximized by following the superstitions of the most successful DX witch doctors. So it was refreshing to read (in *Experimental Wireless and Wireless Engineer*, Dec. 1924) a letter from E. H. Robinson demolishing one of these fetishes, broadly-tuned c.w. transmissions', and ending 'I should like to suggest that the many amateurs who despise theory as not in agreement with practice should inquire more closely into their actual knowledge of theory and the precision with which they observe practical facts'—advice that ought to be pasted in large letters on the walls of many labs as well as shacks.

In fairness to amateurs we must remember the other side of the coin: the professionals relied on their theory that short waves were no good for long-distance communication, so they allocated them to amateurs. They regretted their generosity when the amateurs opened up communication with the antipodes on a few watts. The old theory had to be replaced by a better one.

The supreme example of the ultimate triumph of theory is, of course, the semiconductor revolution. So long as work in this field was guided by practical people it was restricted mainly to fiddling with the catwhisker to find a sensitive spot on the chip of natural galena crystal for better reception of 2LO. That, and perhaps copper oxide rectifiers in the charger needed to keep the l.t. battery fit to heat the bright-emitter valves which were displacing crystals. How vastly different since the physicists stepped in with their Fermi levels and things!

Although examples can be quoted of progress being made with the aid of theories afterwards proved to be wrong, in general it is helpful to pick the right theory. The most entertaining situations arise when theories clash. When Galileo was threatened with dire penalties unless he recanted his theory that the earth moves round the sun (instead of being the fixed centre of the universe) he did so, but is said to have muttered under his breath

'and yet it moves'; just as in my young days we inaudibly added *I don't think* to the declarations of good intent forced on us by parent or teacher. The classic example of a clash of theories in our own field was the great sideband controversy.

Anything of the kind would probably be impossible today. Persons who declare that the earth is flat stand no chance of raising a furore. But around 1930 a world-famous authority in the field of radio, Sir Ambrose ('the valve, invented by me') Fleming, F.R.S., declared—and stuck to it—that sidebands were imaginary; a mere mathematical fiction. This was of more than academic importance. As W. T. Cocking mentions elsewhere in this issue, a new and supposedly revolutionary type of receiver (the Stenode) was claimed to be proof of the wrongness of orthodox sideband theory. And the future of television appeared to hang on it.

Not even the caustic wit of G. W. O. Howe succeeded altogether in dispelling doubts, and in the end a Government-sponsored committee was set up to investigate. Its findings, Special Report No. 12 of the Radio Research Board, published by H. M. Stationery Office in 1932, coldly and relentlessly re-established orthodox theory.

Although hardly such a *cause celebre*, the biggest correspondence I ever had on any article arose from one on a very similar subject, under the title 'Fourier—Fact or Fiction?' (*W.W.* Sept. 1955). Here again the theory is that a current (or other variable) which in practice is generated as a whole, is nevertheless equivalent to a number (if necessary, infinite!) or harmonic components of constant amplitude. Coming 25 years later, this controversy revealed hardly any doubt about the reality of the harmonic components, but centred on a rather more subtle question posed in the article. Granted that there are good practical grounds for the unique place of sine waves in theory of this kind. Could some waveform other than sine wave be regarded as basic and an entirely different but valid theory be erected on that basis? One correspondent declared this discussion to be 'singularly fruitless'. So it may have been in this particular case, but one of the most fruitful mental exercises is to do just what was suggested: question our assumptions. This is a hard mental exercise, so is seldom done. Progress is made by doing it. We tend, lazily, to cherish the concepts we have been brought up on and which have served us well, until we are convinced there can be no reasonable alternative.

I have become only too familiar with this kind of inertia in connection with phasor diagrams. This valuable aid to visualizing the workings of electronic circuits is rejected by all but a few because it looks like an unfamiliar—and therefore difficult—variety of the 'vector' diagrams that we know so well. These we have already found to be almost useless for electronic circuits and no help in understanding ordinary electrical a.c. circuits unless those circuits are simple

enough to be ... valuable without. These old 'vector' diagrams must bear a lot of the responsibility for the widespread teaching (even in textbooks) that there is a phase reversal between primary and secondary of a transformer. (With incredible self-control I'm going to say no more here about my alternative!)

Just after the last World War another controversy raged, about valve equivalent circuits. This time even G. W. O. Howe seemed to lack his usual gift for clearing away the red herrings and revealing the essential, which in this case is that the valve and its equivalent circuit are two quite different things. So the source of anode d.c., necessary for one, has no place in the other. The reference direction of output signal current in the equivalent circuit must be decided without reference to that d.c. source. A pity that muddled thinking on this plagues present-day transistor conventions.

More unnecessary confusion was introduced into valve amplifier theory by the work of Bode admittedly classic, but misguided) on negative feedback in that the symbol  $\mu$  (a universally accepted standard symbol for voltage amplification factor) was used for actual voltage amplification; which is better denoted by  $A$ . The devil of it was that  $\mu$  is correct in the formula for output impedance with feedback, but not in the others, for gain, distortion reduction, etc.

During the last War I was shocked to find many radar instructors teaching that when (say) a positive-going input signal is applied to a *CR* circuit, as shown in Fig. 1, the output also goes positive *because of the charging of C*. In fact, of course, any charging or discharging of *C* appears only as *distortion* of the signal at the output.

Talking about distortion, I would like to mention a variety which I named scale distortion (*W.W.* 24th Sept. 1937, p.318; also Nov. 1948, p.392). When thrown into any assembly of audio experts this term used to act like a petrol bomb in Bogside. Guaranteed instant controversy. I don't know whether it still works. The main point was that unless an audio scene is reproduced at the same intensity at the listener's ear as the original, what he hears is a distorted version, in that the tonal balance (especially the proportion of bass) is upset.

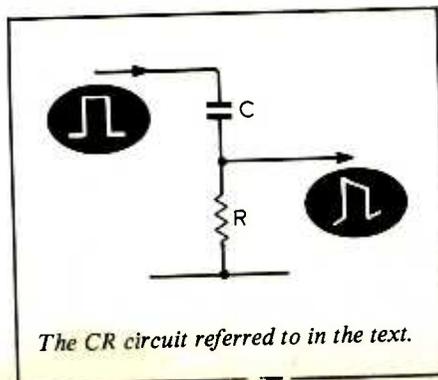
In this statement, only the word 'distortion' could be seriously questioned. Often it was denied that it could be distortion, since the same effect was noted if one moved away from the original

and, or nearer to it. Not quite true, for certain spatial effects would change too; but ignoring them I feel that if the objector had paid for the best seat, acoustically speaking, for a piano recital, and was obliged to sit half-way along the corridor, with the door open, or close against the piano, he might well use an even ruder word than distortion'. The comparison should be with the original heard *in a favourable position*. The other main argument concerned what, if anything, one could do about it when the reproduced intensity was unavoidably not the same as the original. Attempts to 'compensate' by having a tone control linked with the volume control were, in my view, misconceived.

Controversy tends to wax especially hot when, as in this example, *subjective* aspects are involved. Even a stone deaf man can measure sound *intensities*, but neither he nor anyone else is likely to agree about *loudness*; still less, unpleasantness. And all our work ultimately does involve subjective perception, rather than purely physical

What of the future? Since in all our theories we use concepts (such as fields) which of their nature cannot be said to be absolutely true or false (Essays in Electronics (Iliffe), Chap. 1.) there may come a day when it will be found more convenient to use some alternative concepts. But human inertia tends to keep established concepts established. Then of course we can look forward to the discovery of new phenomena, possibly calling for new concepts. But the way things are going I would expect the greatest developments to concern the aforementioned mysterious link between objective phenomena and subjective perception. As a result of new theories of perception we may well be able to improve our equipment by discovering more precisely the physical factors that yield desired or undesired subjective effects. The Haas effect, which decides whether voice reinforcement appears to come from the actual speaker or not, is an example of what has been done along these lines.

To realize how difficult is investigation in the subjective, compare the most sophisticated computer that exists or can be imagined with a human being of very limited attainment—say, an average young child. Even a child has aesthetic likes and dislikes, outside the capacity of the computer. Why? Can strictly scientific investigation tell, or is it for ever beyond its scope? Is it because human beings are not just machines?



The CR circuit referred to in the text.

# The World of Amateur Radio 1911—1971

By Pat Hawker, G3VA

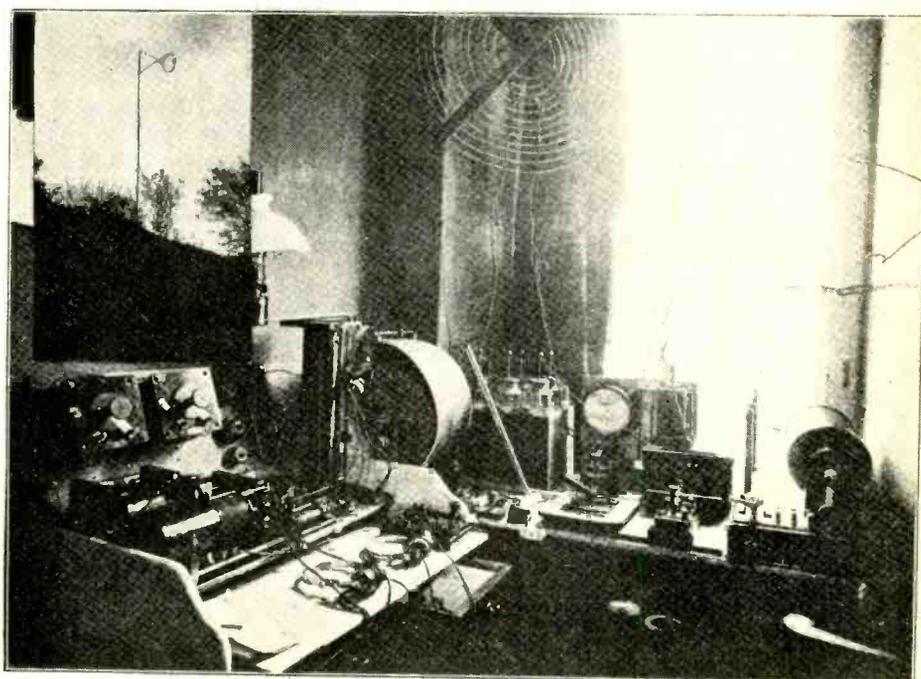
From a few thousand to half-a-million transmitting amateurs . . . from spark to s.s.b. . . . from a few miles to moonbounce . . . from coherers and crystal detectors to m.o.s.f.e.t.s and integrated circuits . . . from jiggers, slide tuning coils, high-speed rotary gaps, motor ignition coils and Wimhurst machines to transceivers and transistors . . . from the bright emitter valves of the twenties and the wide open spaces of "200 metres and down" to the current pressures on the narrow frequency assignments . . . from causing broadcast interference on the old 440-metre band to television interference, the dominating problem of the past few decades. Yet always the same interest in extending working ranges whether on 200 metres or 2400 MHz: the same problems of how to improve receiver selectivity and sensitivity—and almost always the same fears that other interests and authorities may wish to limit the scope of amateur communication.

Historically, amateur or 'experimental' radio reaches back more than 60 years—indeed it began as soon as the results of Marconi and the other pioneers started to reach outwards from purely scientific circles. In the U.K., its official beginning was determined by the Wireless Telegraphy Act, 1904, with its provisions for the use of 'wireless telegraphy for experimental purposes'—though the hobby remained unregulated in the U.S.A. until December, 1912, when the amateurs there were banished to the 'useless' and unwanted regions of '200 metres and below'.

In the pages of 60 years of *Wireless World* one can trace the major events which make up this durable interest. You can find descriptions of pre-1914 amateur stations, including W. K. Alford's station TXK the forerunner of his present call G2DX; the formation in 1913 of the Wireless Society of London which, in 1922, became the R.S.G.B. and for which *Wireless World* was the official journal until February, 1925.

## Early days

It was in the years from 1911 that amateur radio emerged as an activity distinct from the early dabbling in wireless telegraphy—then, as now, with strong links with similar activity in the U.S.A.



This illustration accompanied a description of W. K. Alford's station in the October 1914 issue of *Wireless World*.

Yet the links with modern operating were by no means tenuous. *Wireless World* in 1914 reported an attempt by the Wireless Society of London to put an end to 'jamming' that reads as though it could have been written yesterday. Among the 13 rules were: transmitter to be sharply tuned; receiver to be as selective and sensitive as possible. Those early amateurs were advised: 'do not transmit at less than 12 w.p.m.; listen carefully before calling a station; never carry out testing work with the aerial on that can be done equally well with it off; always use minimum power; refrain from answering a station that is calling some other station; listen in after finishing a conversation to see whether anyone is waiting to call you . . .'

And the clubs were forming. A Junior Wireless Club (later to become the Radio Club of America) had been formed in New York in 1909 at the same time as Hugo Gernsback began publishing *Modern Electrics* with a large radio content. In 1911, the year *W.W.* began as *The Marconigraph*, five enthusiasts formed the Derby Wireless Club, which—as the Derby and District Amateur Radio

Society—is currently engaged in marking its own 60th anniversary. The club met weekly, possessed a library, carried out experiments—one of these was to use two buckets as a variable capacitor; one suspended from a pulley by a cord to raise and lower it into the other! Those Derby enthusiasts soon decided that 'the secret of successful reception is a high aerial and a good pair of headphones'.

In 1912, the first British Commonwealth 'national' radio society was formed—the Wireless Institute of Australia (still the society representing Australian amateurs) and it was reported that some 400 experimenters in New South Wales were covering distances up to 30 miles with converted  $\frac{1}{2}$ -inch motor ignition spark coils.

In Britain, by 1914, 990 of the old three-letter experimental calls had been issued; there was even, in 1913, a 'directory of stations' published by Gamage's which had emerged as one of the main suppliers of equipment to amateur experimenters. Meanwhile the hobby was going great guns in the United States, with A.R.R.L. formed in 1914, but then as only

one of a number of groups catering for amateurs. After the close-down of British stations in August, 1914, the Americans continued active until 1917 with as many as 6000 stations, occasionally working distances over 1000 miles. One of the early American amateurs was none other than Edwin Howard Armstrong whose regenerative receiver, used in his amateur station, was the first of this brilliant engineer's many inventions.

## Broadcasting and short waves

After World War I came a long struggle to re-establish amateur radio transmitting (and even receiving which had equally been banned during the war). In this struggle *Wireless World* played a major role. As early as March 1919 an editorial, with the support of Marconi, Fleming and Eccles, made a strong plea for the removal of restrictions; Marconi wrote 'I consider that the existence of a body of independent and often enthusiastic amateurs constitutes a valuable asset towards the further development of wireless telegraphy.'

But it was late in 1920 before transmitting licences were again issued with call signs of current form (except for the absence of official international prefixes until 1928) for 1000 metres (later changed to 440 metres) and below 180 metres. By this time word was trickling through of long distances being covered by American amateur spark and valve (c.w.) stations: 500, 1000 and even 2000 miles on wavelengths of about 200 to 230 metres.

Soon plans were launched for transatlantic tests, with the British listeners organized by Philip Coursey, 2JK, then research editor of *Wireless World*. The first tests in February, 1921, proved a failure, but new tests were planned for the following winter. Paul Godley, American 2ZE, brought over receivers which he set up at Ardrossan, Scotland. During the December tests he logged 27 American and one Canadian station, but this time several British listeners also heard transatlantic signals. On 11th December, 1921, the special Radio Club of America station 1BCG sent the first complete amateur message across the Atlantic—one of the signatories was Armstrong. These tests finally settled the spark versus c.w. controversy, overwhelmingly in favour of valve transmitters.

During his stay in England Paul Godley lectured to the Wireless Society of London saying 'one has far greater hopes of being able to travel greater distances on shorter wavelengths than on higher wavelengths'. This was a hint of things to come—the historic pioneering of short waves by amateurs. The breakthrough came in late November 1923 when Leon Deloy, French 8AB, worked the A.R.R.L. station of Fred Schnell, 1MO, on about 100 metres, initiating a mighty rush to short waves. Jack Partridge, 2KF, was the first British station to work the Americans, on 8th December, 1923. In January, 1924, E. J. Simmonds, 2OD, made contact with

the States with an input of only about 30 watts. In the autumn Cecil Goyder, a 16-year-old schoolboy at Mill Hill, and Gerald Marcuse, 2NM, made contact with Frank Bell, Z4AA, in New Zealand. It was feats such as these that really woke communication companies to the possibilities of the short waves they had previously scorned.

Not all amateurs in the 'twenties were interested in long-distance working. In 1920-21, with no regular broadcasting in the U.K., British amateurs began putting out gramophone records and even live concerts—sometimes with official blessing, sometimes without it. The Wireless Society organized a massive petition, signed by 63 clubs representing 30,000 enthusiasts, asking for regular broadcasting. Amateurs can also claim to have initiated Empire broadcasting when Gerald Marcuse, 2NM, began a series of transmissions on about 30 metres, well in advance of the first B.B.C. station G5SW.

But the coming of broadcasting soon brought problems to the amateurs, and 440 metres was lost. By then Hugh Pocock had been writing in *Wireless World* of the extent that petty rivalry was threatening to split the ranks of amateur enthusiasts. But the biggest threat was in the new regulations of April 1924 when the authorities attempted to impose a total ban on all casual international working by British stations. *Wireless World* immediately offered to place £500 at the disposal of the R.S.G.B. to allow a test case to be argued in the Courts. Fortunately the authorities gradually gave way (although the ban on the use of 'CQ' remained until 1946).

## Later days

In the years that followed, amateur radio became established as something quite distinct from broadcasting and turned more and more to h.f. and v.h.f. Strictly speaking all licences in the U.K. were 'experimental' as the first 'amateur' licences were not issued until 1946. But, in practice, in the 'thirties the hobby was recognized, even if newcomers were restricted to 10 watts and 1.8, 7 and 14 MHz.

Frequencies continued upwards. Jimmy Matthews, G6LL, (still active) made the first transatlantic contact on 28 MHz in October, 1928, but the band went dead for some years. About the same time, amateurs began to explore frequencies around 56 and 112 MHz, then wide-open and unused. In the 'thirties came the annual contests and the operating certificates. Equipment which until then had been virtually all home-built began to be purchased. The superhet communications receiver began to displace the 0-V-1\* and 1-V-1 receivers. Names such as Eddystone, HRO, Hammarlund and Hallicrafters were heard increasingly. The

crystal filter, developed for h.f. by Lamb of A.R.R.L., exploited a British development for stenode reception.

By now crystal control had replaced the old t.p.t.g. and master oscillators (and was later to be partially replaced by v.f.o.). The 'Zepp' was the popular aerial of the 'thirties. Near Chicago, Grote Reber, W9GFZ, was making the first-ever radio telescope to plot radio sources in the sky—while Dennis Heightman, G6DH, identified solar hiss. A.M. stations on 14 and 28 MHz made intercontinental phone working a daily occurrence.

On 31st August, 1939, all British experimental licences were withdrawn and amateur activity, as such, ceased until January, 1946. Many amateurs used their operating skills in the Services—and incidentally were responsible for many of the most successful feats of both British and German intelligence! Indeed, interest in the hobby did not fade away, and the R.S.G.B. actually increased membership from 3800 to 9600 during this period.

In February, 1941, a controversy arose following the publication of an article 'The future of amateur radio' in *Wireless World* (and which I can now reveal was written by me), as to the form that post-war licences should take. Many of the ideas put forward during this controversy were later adopted by the Post Office when, in January 1946, the first-ever U.K. amateur licences were issued, though a number of 'incentive' elements of this licence were later to be abandoned.

But the basic terms of the 1946 licence continue through several revisions to govern the hobby in this country, though the introduction in 1964 of Class B (phone-only, v.h.f.-only) licences poses the likelihood of further erosion of interest in c.w. telegraphy, which to some of us continues to be 'inescapably the basic form of amateur communication'.

By the 1950s, the first steps towards the now dominant s.s.b. mode of operation had been taken, 144 MHz had displaced the old 56 MHz band, 420 MHz had been opened to amateurs (including the new interest in amateur television), rotary Yagi and Quad aerials had become popular, amateurs had discovered transequatorial propagation. Two further decades have brought the transceiver and the semiconductor revolution, but the hobby has changed remarkably little in its essentials.

One might ask what is the appeal of amateur radio that attracts recruits from both inside and outside electronics and succeeds in retaining their interest not just for the average 10 years or so, but for 30, 40, 50 and even 60 years? Its recruits have included such giants as Armstrong, Ryle, Kraus, Campbell-Swinton, Fink, Terman, Henney, Collins, Eckersley, Reber, Villard, Crosby, Beverage . . . and also 'outsiders' such as Barry Goldwater, Brian Rix and King Hussein.

Perhaps *Wireless World* summed it all up when, in 1914, it quoted Tennyson:  
I hear at times a sentinel  
Who moves about from place to place  
And whispers to the worlds of space  
In the deep night that all is well.

\* Old method of classifying stages in a receiver; first numeral, number of r.f. stages; letter 'V', detector; second numeral, number of a.f. stages.

# F.M. Stereo Tuner

## High-performance design for home construction

by L. Nelson-Jones, F.I.E.R.E.

In recent years there have been a number of developments in the components field, particularly in semiconductor devices, that have led to great improvements in the design possibilities for f.m. broadcast receivers. In particular these have been the advent of the dual-gate m.o.s.f.e.t., integrated-circuit i.f. amplifiers and demodulators, ceramic filters and improved variable-capacitance diodes. This two-part article describes an f.m. tuner design using these devices, discusses the advantages of the devices and gives constructional and alignment details. It does not attempt to be all-embracing and there will doubtless be some who disagree with the author's views of the current scene. It is hoped however that they do show some ways in which f.m. tuner design is currently evolving.

The work is the result of many months of design and measurement on five

prototypes, so that results are not based on a one-off, and should be reproducible by readers who wish to copy the design. The receiver was designed to achieve in a relatively simple way a performance equal to the better examples of the commercial models available, but at a much lower price. (Total material cost comes out at about £11.) Comparison with the figures given in a recent *Wireless World* survey of commercial tuners (September 1970) suggest this aim has been achieved. The performance of the tuner under normal conditions of use has been excellent. One of the units is in use in Blandford Forum in Dorset—very much a fringe area—and gives noise-free reception from the Isle-of-Wight transmitters, including the new local station Radio Solent.

The design for the front-end of the f.m. receiver is shown in Fig. 1. Both r.f. amplifier and mixer stages use dual-gate

f.e.t.s with gate protection diodes. In the r.f. stage the upper gate is decoupled and acts as a screen between drain and gate 1, much as the  $g_2$  electrode of a thermionic valve does. In the mixer stage this second gate is used as the injection point for the local oscillator voltage. There is not the same need for a screen between drain and gate 1 in a mixer stage as the drain load is not tuned to either the signal or oscillator frequencies. There is therefore little or no gain at signal frequencies to cause oscillation provided care is taken with the layout, particularly the length and placing of leads.

The magnitude of the local oscillator injection at gate 2 will affect the mixer gain and the spurious signal response characteristics of the mixer stage. This local oscillator voltage will be higher than in transistor tuners using bipolar devices by up to an order of magnitude and for the circuit conditions used a value of

Fig.1. Front-end of receiver using dual-gate f.e.t.s.

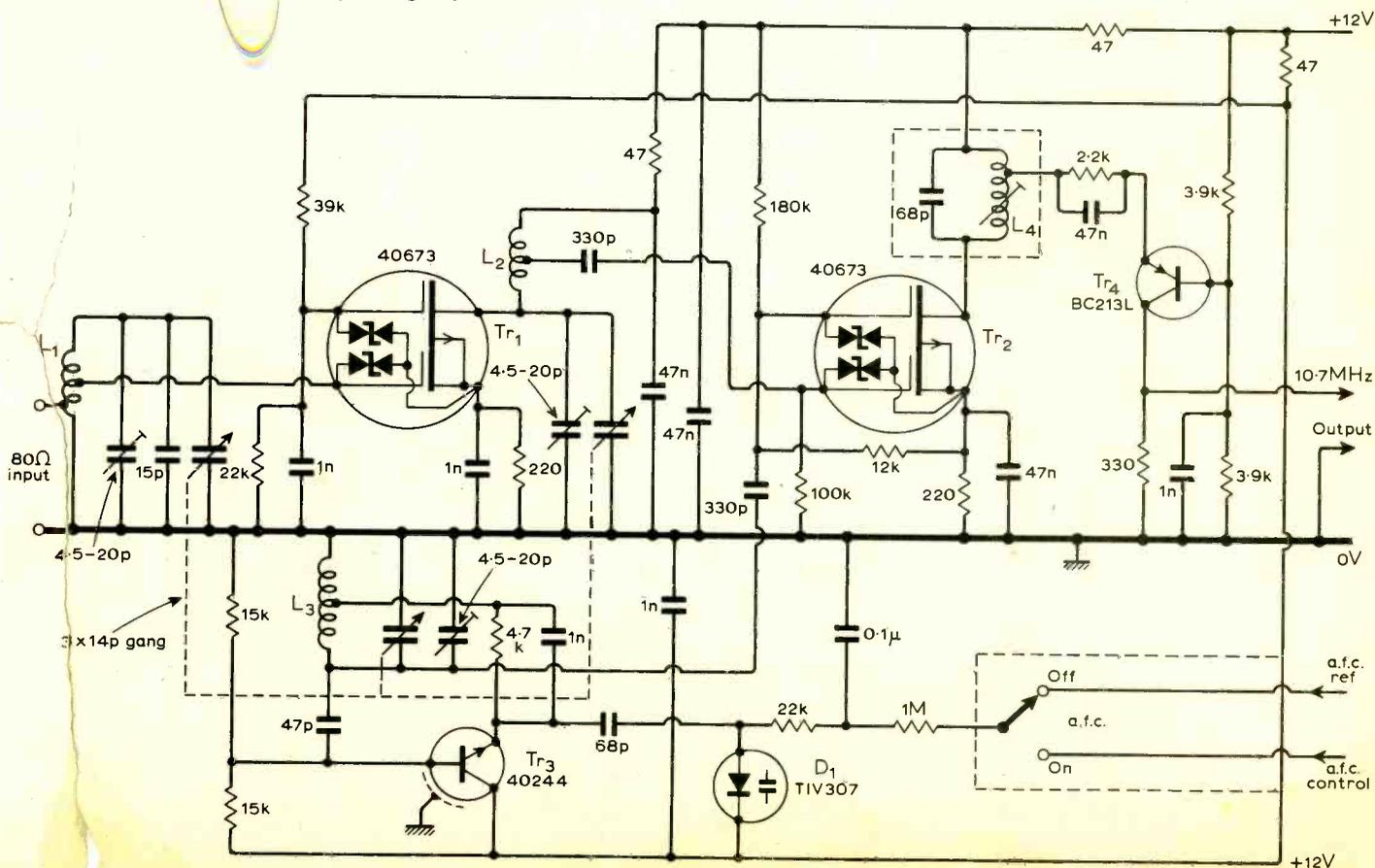
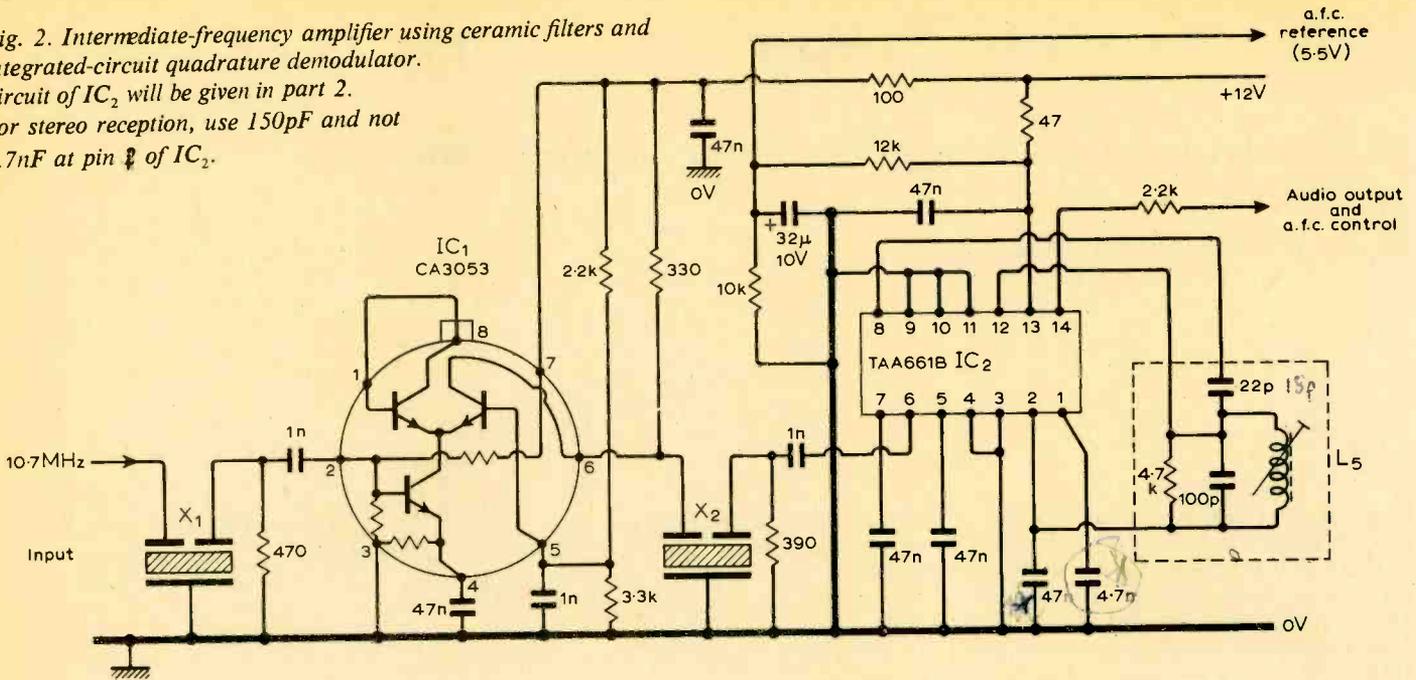


Fig. 2. Intermediate-frequency amplifier using ceramic filters and integrated-circuit quadrature demodulator. Circuit of IC<sub>2</sub> will be given in part 2. For stereo reception, use 150pF and not 4.7nF at pin 4 of IC<sub>2</sub>.



500mV r.m.s. gives a reasonable mixer gain without too high a spurious signal response. In fact higher levels have been used without great trouble from spurious signals. Far greater trouble can be caused by lack of screening, leading to i.f. harmonics being picked up by the front end—especially with the high sensitivity of this tuner and, because of its small size, the close proximity of the front-end and the i.f. amplifier.

The oscillator is a conventional Hartley circuit with the ground point moved to give a grounded-collector design. There is no particular advantage to be gained in using an f.e.t. in this stage so that the cheaper bipolar device is preferred. Automatic frequency control is applied by the variable-capacitance diode D<sub>1</sub> coupled to the emitter of the oscillator stage. A resistor of 22 kΩ prevents the decoupling capacitor of 0.1μF from shorting out the oscillator voltage. The 0.1-μF capacitor together with the 1 megohm resistor form a low-pass filter to prevent audio voltages in the a.f.c. voltage from reducing the modulation of the carrier by audio frequency modulation of the local oscillator.

The a.f.c. can be switched out of operation by connecting the diode to a constant reference voltage from the i.f. amplifier. Diode D<sub>1</sub> is returned to the 12-volt supply line of the oscillator so that the a.f.c. control voltage changes the diode reverse voltage in the correct direction to reduce any oscillator drift. An increase in local oscillator frequency increases the intermediate frequency, which in turn leads to a rise in the output potential of the demodulator of the i.f. integrated circuit IC<sub>2</sub> (Fig. 2). As the diode is connected to the +12V supply, this increase reduces the reverse bias across D<sub>1</sub>, increasing the diode capacitance and reducing the local oscillator frequency to correct its drift.

The mixer has a grounded-base stage feeding the 330-ohm resistor needed to correctly terminate the first filter unit X<sub>1</sub>

(Fig. 2). This resistor also makes a convenient low-impedance output point from the front-end. A cheap p-n-p bipolar device is more than adequate for this position, because in a grounded-base configuration the requirements in respect of high frequency or noise performance are not stringent. The working Q of the tuned circuit is less than 20 so that tuning is not critical, and it is set to maximize gain in the usual way.

The supply for gate 2 of the r.f. stage is derived from the decoupled oscillator supply rather than from the top of L<sub>2</sub>. This is brought about purely by layout convenience on the printed wiring board and, as gate 2 is additionally decoupled, has no effect on the performance.

**I.F. amplifier**

Two ceramic filter units X<sub>1</sub> & X<sub>2</sub> are separated by a buffer amplifier (IC<sub>1</sub>) of moderate gain (about 20dB). The reason for this moderate gain is that it is desirable to place the filters as early as possible in the i.f. amplifier so that as successive stages limit with increasing signal level there is no change in bandwidth. This would be fully achieved if the whole of

the i.f. gain were after the filter, and provided the mixer did not limit.

In practice it is not possible to achieve this ideal, but the compromise of using only moderate gain before the second filter unit is a reasonable one and does not give rise to any undue increase in bandwidth over the normal range of signal levels. The performance obtained is a great improvement over normal bipolar i.f. amplifiers using discrete components with several double-tuned i.f. transformers. In such an amplifier the selectivity of the transformers is gradually lost starting at the output end as successive stages limit and the overall selectivity can leave a lot to be desired at high signal levels.

The first integrated circuit is a long-tailed pair circuit, used as a cascode amplifier by ignoring one of the top pair of transistors and driving into the long-tail transistor. The input impedance of this stage is suitable for the ceramic filter unit so far as resistance is concerned, but is above the maker's recommendations so far as input capacitance is concerned. For this reason the resistor terminating the filter X<sub>1</sub> is raised to 470 ohms, which compensates for the increased capacitance loading of the stage in restoring the top of the filter characteristic to reasonable flatness.

The load of the cascode stage is a 330-ohm resistor—to drive the filter X<sub>2</sub> from the correct source impedance. This low value results in the stage gain being low, especially when the loading effects of the filter are accounted for, so that although the slope of the cascode stage is around 100mA/V the overall gain from the output of X<sub>1</sub> to the input to X<sub>2</sub> is only a little over 20dB.

The input impedance of the IC<sub>2</sub>, around 2 kΩ, is not very much greater than 330 ohms so that a terminating resistor of 390 ohms is used at the output of X<sub>2</sub>. The value of the feed capacitor 'quadrature' tuned circuit L<sub>5</sub> is larger than the maker's recommended value of 18pF.

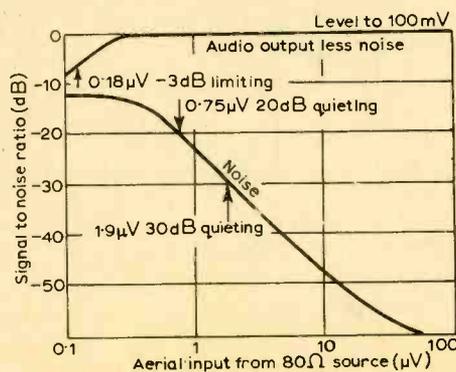


Fig. 3. Graph of low-level performance shows sensitivity of 0.75μV (for ±75kHz) for 20dB quieting. Above 50μV input, signal-to-noise ratio is better than 60dB.



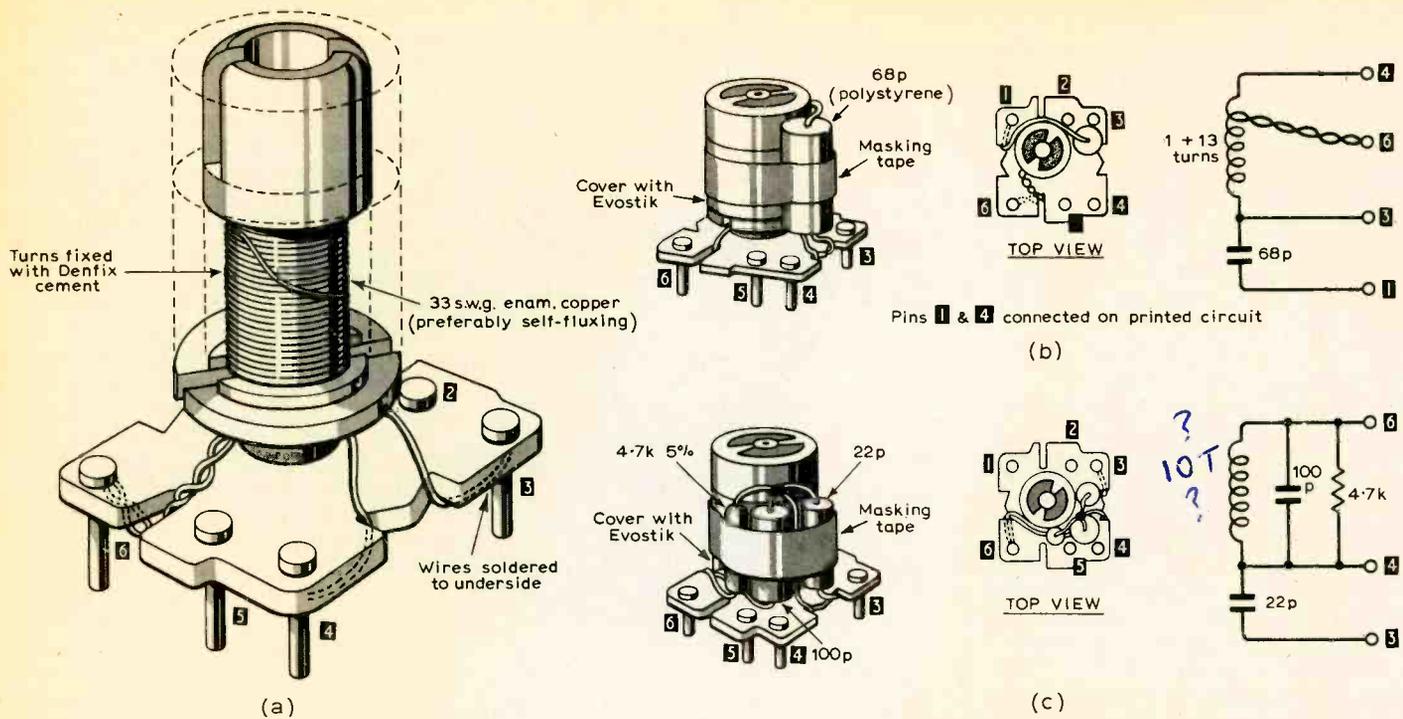


Fig. 6. Two screened i.f. coils are wound with 33-gauge enamelled wire and secured with a little Denfix cement. Capacitors are fixed with masking tape.

the signal-to-noise ratio is better than 60dB, and the 20 and 30dB quieting figures are 0.75 and 1.9µV respectively. Limiting (-3dB) of the demodulated audio signal (single-tone filtered from noise) occurs at only 0.18µV so that at all usable signal levels the i.f. section is limiting.

Unfortunately no signal generator was available which had an output with amplitude modulation-free from f.m., and it has therefore not been possible to check the a.m. rejection of the receiver. However, it is expected that the result will be close to the figure quoted for the i.f. integrated circuit (IC) at moderate signal levels, with an improvement when the first i.f. amplifier limits. The figure quoted for the TAA661B is 40dB at 10mV input, equivalent to around 10µV input at the aerial. Performance is summarized in the table on this page.

**General layout**

The tuner is constructed on a single-sided printed circuit board and is divided into two areas. The front-end and the i.f. amplifier are laid out separately, side by side, on a printed circuit board about 10 x 8 x 5cm overall, and in such a way that they may be separated if desired.

The complete tuner is enclosed in a screened box to cut down on spurious responses due to radiation from the i.f. amplifier, and to reduce local oscillator radiation to a minimum. This screening is especially necessary with this design because of its very high sensitivity.

The dial drive system suggested gives a scale length of 13.7cm with a reasonably linear frequency change over the centre 80% of this scale. A cord drive system is used which has the advantage of retaining the pointer at both ends, thus eliminating the problem of sliding friction at one end

of the pointer, and giving a much smoother drive.

The overall layout of the components is shown in Fig. 4 as seen from the components side of the board, and also in the oblique view.

In construction keep leads short and if possible test all components on a bridge before fitting them on the board, as this can save ruining a good p.c. board should the tuner not work first time. It cannot be emphasized too strongly that such component checks can save much wasted money and temper. It is also vital to check that the components are correctly located, the diode connected with the right polarity, and that there are no breaks or shorts on the "track" of the p.c. board. This latter point is of importance on such a small board with roughly 200 component holes, as tracks are necessarily fine and gaps small. A watchmaker's eyeglass has been the constant companion of the author during the construction and design of this tuner.

**Coil construction**

The r.f. coils are all made from 18-gauge tinned copper wire and are self-supporting. Taps are made by soldering leads of 22-gauge tinned copper direct to the turns of the coils—Fig. 5. The coils were made by winding the wire on a 1/4-in rod such as a drill shank. The wire should be straightened by placing one end of a length in the vice and pulling the other end until the wire stretches very slightly, when the wire will have lost all kinks.

The coil wires should be a firm push fit in the board, if undue strain on the copper foil of the board is to be avoided when adjusting the coils. At all cost avoid the wires being loose in the board before soldering and if necessary apply the

minimum of Araldite epoxy adhesive around the 18-gauge coil wires on the components side of the board after soldering in position. The joints should then be quickly reheated with the soldering iron to cause the Araldite to run into the holes in the board, thus securing the coils rigidly. After using such an adhesive the board must be left in a slightly warm place, e.g. an airing cupboard, for 24 hours to ensure that the adhesive has set hard.

The coils should be mounted on the board in the positions shown in Fig. 4, and with the turns of the coils nearest the board surface 2.5mm clear of the board. In the case of the oscillator coil this must also be 3.5mm clear of the rear face of the tuning capacitor. It is best to adjust the coils before soldering them into the circuit board to minimize subsequent adjustment, and the overall coil lengths given (over the outside of the end turns) are close to the

**Performance**

<b>Sensitivity</b>	
-3dB limiting	0.18µV
20dB quieting	0.75µV
30dB quieting	1.9µV
<b>Spurious response</b>	
image rejection	-70dB
i.f. rejection	-85dB
other unwanted signals	-94dB
<b>Audio output</b>	0.5V r.m.s. for ±75kHz
<b>Capture ratio*</b>	2dB approx.

\* Difficult to measure. There does not appear to be much dependence on signal level provided the signals are reasonably above noise level—a result to be expected from the very low level at which limiting starts. Figures varying from 1 to 4dB were measured at various signal strengths on repeated measurements. In general a signal 10dB below produced no noticeable effect on the demodulated output.

final adjusted lengths required for correct tracking.

An alternative method of mounting is to open up the main coil mounting holes in the board and insert 'eyelets' which are big enough to allow the 18-gauge coil leads to pass through them. The eyelets are riveted into the board so that the strain is removed from the copper track. Overall connection is obtained by soldering the track to the eyelet and to the coil leads. All the main pads for the ends of the coils are large enough to allow this to be done.

The two screened i.f. coils are both constructed on Neosid coils type NS/E3, and both are wound with 33-gauge enamelled wire, preferably of the self-fluxing variety. The two coils are wound as shown in Fig. 6 and the turns secured in place for stability with a minute quantity of Denco Denfix polystyrene cement. (Do not use modelmaker's polystyrene cement as some varieties can have high loss factors.) It is essential to use the least possible quantity as the bobbin of the coils is made from polystyrene loaded with iron dust and is very easily dissolved by this cement. When the cement has dried, place the ferrite sleeve over the coils, ensuring that the leads are well pressed down in the slot at the base of the bobbin so that this ferrite sleeve does not scratch the wires.

Next push on the polythene retaining disc. Secure the ferrite sleeve to the coil former with a smear of Evostik latex-resin contact adhesive around the join between the sleeve and the coil former near the base. When dry connect the capacitors (and the resistor in the case of  $L_5$ ). Tape these components to the former as shown to hold them clear of the coil can. Make sure all leads are well clear of the can when this is slipped over the coil. Next fix the core into the coil. The ferrite core cuts its own thread into the polythene top retainer. Set the top of the core level with the top of the polythene retainer—Fig. 6b. This should be close to the final adjustment position. In the case of  $L_4$  the capacitor is connected to the coil on one side via the printed circuit, which connects pins 1 and 4, thus placing the 68-pF capacitor across the coil.

**Fitting components to the board**

Due to the small size of the board and components it is absolutely essential to use a soldering iron with a small tip which is adequately hot and clean. Lead lengths must be kept short and the components close to the board. This is especially important with the ceramic disc decoupling capacitors (1 and 47nF types). The transistors should be pushed down onto the board until the body is 2.5mm above the board; pushing the transistors closer than this will strain the leads unnecessarily. This rule applies also to  $IC_1$ . The second i.c. should be placed down on the board until the shoulders on each lead contact the board; the body of the i.c. will then be just clear of the board.

There will be some difficulty in locating the polarity of the diode due to the small size of this device. If in doubt check it with

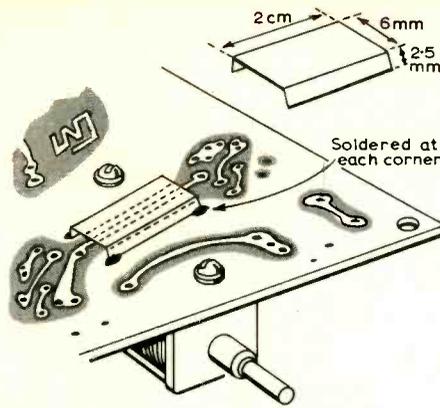


Fig. 7. Metal bridge—covering lead from  $Tr_1$  drain to its tuned circuit—under the tuning capacitor is essential to maintain stability.

an ohmmeter on the ohms  $\times$  1 range, when the cathode will be the one connected to the positive lead of the multimeter when the meter shows conduction. (On a multimeter the polarity of the leads on the resistance ranges is the opposite to that shown on the meter panel so that the red positive lead is negative, and when connected to the cathode results in the diode being forward biased.)

There are two wire links on the board, one on either side of  $L_5$  in the i.f. section. These links may be either 22-gauge tinned copper or 1-024 p.v.c. covered wire. Connections to the 3-gang capacitor are by similar wires. The capacitor is secured to the board by two 6BA screws not longer than 4.5mm of thread.

The link from the r.f. to i.f. sections is by a twisted pair of 1-024 p.v.c. insulated wire as shown in Fig. 4 and in the oblique view. Take care to see that this is correctly connected, i.e. the live lead of the pair connects between the collector of  $Tr_4$  and the input to  $X_1$ .

The two screened coils  $L_4$  and  $L_5$  are soldered into circuit after being pushed

well down on the circuit board so that when the can is placed over the coil it just rests on the polythene retainer at the top of the coil, while also just contacting the p.c. board. The can is put on the coil after soldering the coil into the board, and the two can tags are then also soldered to the board.

There is one component not shown on the circuit because it is not a circuit component. This is a 'bridge' across the lead from the drain of  $Tr_1$  to its tuned circuit under the tuning capacitor. This bridge continues the earth plane as well as screening the lead. It is essential to use this bridge to maintain stability in the r.f. amplifier. The bridge is necessary because of the layout limitations set by the capacitor having its connections on only one side of the body, and the need to keep the coils well spaced to obtain good stability and keep oscillator radiation low.

The dimensions and location of this bridge are shown in Fig. 7. The bridge may be made of any metal that will not corrode but will solder. Tinplate was used in the original units. Take care not to short the wire to the centre-section stator of the variable capacitor at the end of the bridge nearest to  $L_2$ .

If the ceramic filters need removing from the board, take care not to apply pressure when applying heat to the connections, otherwise the component will be damaged. Remove solder first with a desoldering tool or with copper braid.

In the photograph the alternative type of oscillator transistor is shown. If this type is used then the fourth connection on the transistor won't exist—used to earth the can on the TO-72 type specified (40244). Only the three connections nearest the 3-gang capacitor will then be used. In addition, an extra lead is shown in the photograph adjacent to the integrated circuit  $IC_2$  that is not shown in Fig. 4. This lead connects to pin 14 of this i.c. to control the a.f.c. However, as the output lead (via the 2.2-kohm resistor) will

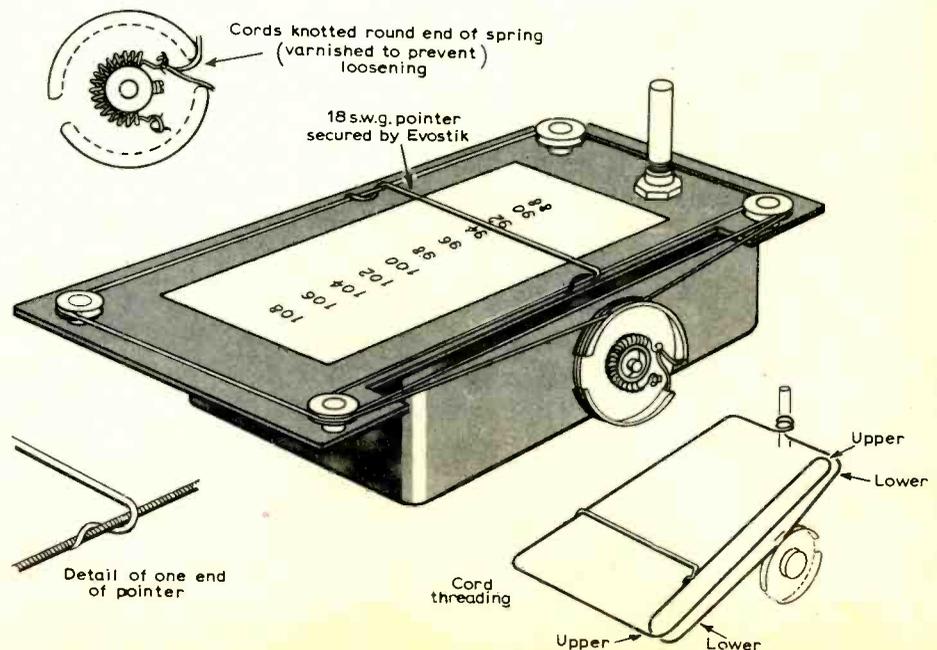


Fig. 8. Suggested cord system eliminates pointer friction and can be mounted in any plane.

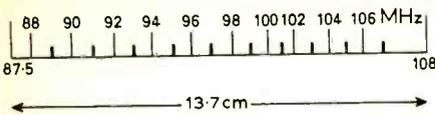


Fig. 9. Typical scale graduations for the band 87.5 to 108 MHz.

normally have a d.c. blocking capacitor in series, it will be at the same potential as pin 14, and is therefore suitable for the same purpose.

### Screening of the tuner

It is essential to screen the tuner to avoid instability in the i.f. section due to pick-up, particularly from the output lead. Although the de-emphasis capacitor removes most of the 10.7-MHz signal and its harmonics, the output lead can still have sufficient of these signals present to cause spurious whistles when tuning if this lead is anywhere near the r.f. section. A great improvement results from connecting a capacitor of 470pF from the output to ground which, with the 2.2k  $\Omega$  series resistor, removes these high-frequency signals sufficiently to make the position of this lead less of a problem. (In stereo applications the capacitor used will probably need reducing to 100-200pF to avoid attenuating the stereo switching waveform unduly, and if a long screened lead is used for the output the capacitance of this lead should be enough by itself.)

Prototypes were fitted into an ITT die-cast case with internal dimensions 12  $\times$  9.5  $\times$  2.5cm and is a tight fit with regard to height. Connections to the tuner should be made through insulated feed-through terminals close to the board so that all leads are as short as possible inside the screened box. A slot will have to be cut in the side of the box to allow the tuning capacitor shaft to pass through. The author found it easiest to fit the tuner board to the lid of the box, and to use the box as a cover, with holes drilled in line with the trimmer capacitors and  $L_4$  and  $L_5$  to enable final alignment with the box closed.

The board is mounted by four 6BA screws as shown in Fig. 4 and must be spaced about 5-8mm from the surface of the screened box, to prevent the track-side of the board from shorting on the box. Extra nuts or spacers may be used to achieve this spacing.

### Dial drive system

Fig. 8 shows the layout of the suggested cord drive system which eliminates the problem of pointer friction, and is suitable for mounting in any plane. The parts required are made by the manufacturers of the 3-gang capacitor with the exception of the pointer, made of 18-gauge tinned copper, or similar, and the cord. Typical scale graduations for the 87.5 to 108-MHz band are shown to scale in Fig. 9.

To be concluded with a discussion of devices used and alignment methods.

### Parts list

Set of parts is available from Integrex Ltd, P.O. Box 45, Derby DE1 1TW.

#### Inductors

See illustrations and text for winding details  
 $L_4$  &  $L_5$  inductor assemblies are Neosid NS/E3

#### Dial components (all Jackson)

$\frac{1}{2}$ -in brass pulleys, type 4879 (4 off)  
 $\frac{1}{2}$ -in brass pivots, type 4539 (4 off)  
brass spacers, type 4880 (4 off)  
type 'G' drive spindle, type 5080  
drive drum 2.5cm dia. 4-mm bore

#### Variable capacitors

3  $\times$  14pF tuning capacitor, part no.5560/3/14 (Jackson)  
4.5 to 20pF trimmer (3 off) (Piher make from Henry's Radio or Rosenthal type STSE-7, N750 from Radio Resistor Co. or type 7S-Triko 02 from Steatite Insulations)

#### Fixed capacitors

1nF disc ceramics, 50V, 1-cm mounting centres (10 off)  
0.1 $\mu$ F, 16V (Mullard type C280)  
32 $\mu$ F, 10V (Mullard type C426)  
22, 68 (2 off) & 100pF, 160V, polystyrene  
15, 47 & (2 off) 330pF ceramic tubular or disc, or polystyrene mounting centres 1cm  
4.7nF miniature tubular polystyrene 1.65-cm mounting centres. Use 150pF for stereo reception

4.7nF disc ceramics, 1cm,  
k7nF 8off-

#### Resistors

Miniature carbon film type,  $\frac{1}{8}$  watt  $\pm$  5% tolerance (Mullard)

#### Active devices

40673 (RCA 2 off). In mixer stage, lack of gate protection diodes may be acceptable, in which case types 40604 or 3N141 can be used. If the risk of not using diodes is acceptable for r.f. stage, types 40603 or 3N140 can be used. N.B.: retain protective spring until power is applied  
40244 (RCA), alternatively TI409 (TO-92) —now available as TIS64 (TO-18), Texas  
BC213L (Texas) or BCY70  
CA3053, 3028A or 3028B (RCA)  
TAA661B (SGS)  
TIV307 (Texas)

#### Also needed are

printed circuit board (available drilled, solder coated and with component locations)  
ceramic filters type FM-4 (Vernitron). Order as pair with same colour coding (orange-10.625 MHz, yellow-10.6625 MHz, green-10.700MHz, blue-10.7375MHz, violet-10.775MHz)  
trimming tool for  $L_4$  and  $L_5$  cores  
nylon cord  
die-cast box  
Denfix cement  
Denfix cement (from Home Radio)  
Evostik latex-resin impact adhesive

## Baird's Video Disc turns again

The video disc is not a new idea, as J. C. G. Gilbert pointed out in his article on the Teldec system last year.\* John Logie Baird recorded video signals from his 30-line television camera on a 78 r.p.m. wax disc as far back as July 1928. Copies were made and sold to the public by Selfridge's store, London, in the early 1930s, the idea being that they should be played from an electric gramophone into the Nipkow-disc Baird Televisor of the time.

Recently *Wireless World* was able to examine one of these discs, and see the kind of pictures it produced on a Televisor, at a demonstration put on by the I.T.A. at the television museum in its London headquarters. The disc, which looks like a 10-inch black gramophone record with a red label, was acquired from Mr. G. Diment of Herne Hill, London, who bought it from Selfridge's in 1935 at a price of 7 shillings. On the label are the words 'Recorded Television Record No. 1, Speed 78, Scanning Speed 750, Lines 30, For Private Use Only'. Because the disc is very fragile a magnetic tape recording had been made of its video signals, and it was this taped copy which was played into the Televisor at the demonstration.

The Televisor was a 40-year old model, one of two acquired by the museum and restored to working order by I.T.A. engineers at the Fremont Point transmitter, Jersey. The authenticity of the results was assured by H. J. Barton-Chapple, who worked with Baird, and by P. J. Packman, who built one of the two Televisors in 1928 at Plessey.

All that can be said of the pictures seen is that they were a sequence of patterns, in the characteristic orange light of the Televisor's neon tube. What the patterns depicted was anybody's guess, though we were told they were caricatures of human faces. Certainly, the first video disc cannot be regarded as anything more than a technical curiosity. It is nonetheless a further tribute to Baird's ingenuity and enterprise in the face of the public's indifference at that time.

\*"The Video Disc", *Wireless World*, August 1970, p.377.

### Correction

Peter Blomley, author of the articles 'New approach to class B amplifier design' (February and March issues), tells us  $T_3$  in Fig.1 of the second article should be type 2N3904 and not 2N3905, and that in Fig. 5 the ordinate should be labelled 0.00075%/cm, and not 0.0012%/cm.

# News of the Month

## 'W.W.' amateur radio station

As part of the journal's 60th birthday celebrations we are to operate during April an amateur radio station using the specially assigned call GB3WW. The station is being set up in Dorset House, where we have had our editorial offices for nearly 40 years, and will operate in the amateur h.f. bands.

Our contributor Pat Hawker, G3VA, will be in charge of the station and among those manning the station from time to time will be F. C. Ward (G2CVV) president of the R.S.G.B., D. A. Findlay (G3BZG) general manager of the R.S.G.B., R. S. Roberts (G6NR), B. M. Johnson (G3LOX), J. Brodzki (G3HQX), S. Andrews (G3OGY), D. R. Bowman (G3LUB) and G. M. C. Stone (G3FZL).

The station will comprise a KW2000B transceiver, KW1000 linear amplifier, Eddystone 1830/1 receiver, Heath SB303 receiver and the s.s.b. receiver described by David Bowman in our July-September 1969 issues. The microphone is a Shure 444. A trap dipole is being installed on

the roof of Dorset House about 100 ft above street level.

We are grateful for the co-operation of manufacturers and the Minpostel in enabling us to operate this station. A special QSL card embodying a reproduction of the front cover of this issue is being prepared.

## Electronics industry manpower

For the first time the manpower problems of the electronics industry in the U.K. have been studied in isolation from those of the electrical industry. This has been done by a working group of the electronics 'Little Neddy' (Economic Development Committee), under the chairmanship of Professor G. D. Sims of Southampton University, and some conclusions and

recommendations are presented in a preliminary report 'Qualified Manpower in the Electronics Industry' published by the National Economic Development Office. The conclusions are not exactly startling, the main points being that while the availability of qualified engineers and scientists is satisfactory there is evidence of a serious shortage both of technicians and of graduate production engineers willing to work in the electronics industry.

There seems to emerge from the report a picture of an industry in a confused and mulish state of mind about its own problems. It does not understand what facilities and information are available to assist with the education and use of technological manpower; it does not have adequate statistics and forecasts for manpower requirements; it is not sponsoring students for 'Bosworth' bridging schemes in sufficient numbers to make the courses fully viable; and it does not regard the release of qualified workers at regular intervals for continuing education as normal and desirable.

Among the working group's recommendations is one that the electronics EDC should sponsor a study of manpower utilization in the context of the management of innovation and its consequences. The fact that U.K. industry is well behind the U.S.A. and Japan in the speed and effectiveness with which it can turn an R & D project into a commercially available product is now well known, and has been made painfully apparent by the Rolls-Royce debacle. The working group is convinced that 'a better use of manpower, better management and control of projects, and a more appropriate mix of manpower resources could all make very substantial contributions to shorter lead-times, and thus to more effective innovation'. The EDC has agreed in principle to the proposal and asked the NEDO to start the study.

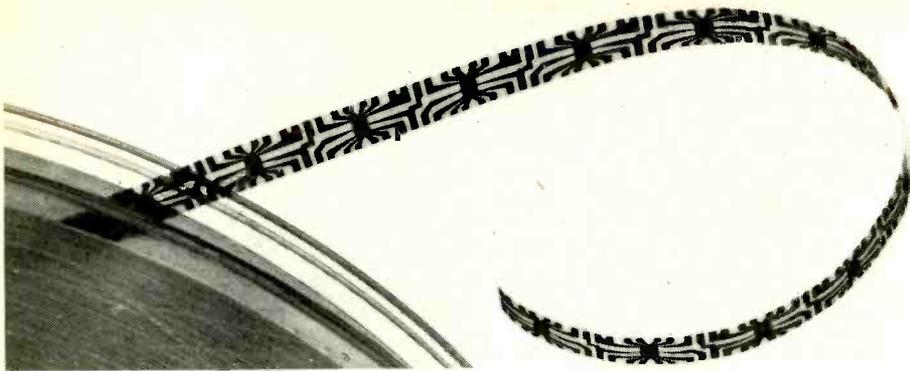


*New sector control consoles in the Mediator air traffic control system are now in full operational use for controlled airspace at the West Drayton centre. Consoles include flicker-free bright displays—made possible by a scan conversion technique—which combine both primary radar information in the form of blips and secondary radar data in the form of digitally generated characters. Secondary radar can provide aircraft identification, route and height information depending on the kind of transponding equipment installed on board aircraft (see 'Progress in air traffic control' page 200).*

## Film and chips

A new method of mounting integrated circuit chips, which looks like being very suitable for use with automatic assembly methods, has been developed by the Philips Research Laboratories, of Eindhoven, Holland. The processed integrated circuit chips are mounted on a long strip of Polymide film and can be stored on reels like recording tape. The Polymide film has a metallic film contact pattern on which the chips are mounted in the same way as 'flip-chips' are mounted in hybrid integrated circuits. After the chips have been mounted the reels of tape holding the chips are treated to improve mechanical strength and are then passivated. During testing faulty chips on the film are marked so that they can be rejected automatically during subsequent circuit building.

The film-mounted chips can be applied



The metalization pattern on the Polymide film can be clearly seen. Integrated circuit chips mounted in this way could easily be used in automatic assembly machines.

to a printed circuit board, or to any other substrate for that matter, or they can be mounted on specially designed headers for conventional insertion into printed circuit boards. When mounted in the header Philips are claiming excellent heat dissipation characteristics. An experimental monolithic audio amplifier with an output of 3W has already been built and operated successfully using the technique.

### Amateur radio history

As mentioned elsewhere in this issue the Derby Wireless Club (now called the Derby and District Amateur Radio Society) was formed in 1911, and, like *W.W.*, is marking its 60th birthday with special events. The first is an exhibition to be held in the Derby Museum and Art Gallery from 3rd to 17th April (excluding Good Friday). The aim of the exhibition is to illustrate the various activities of the society, but space will also be devoted to the history of amateur radio. Exhibits will include items constructed by early members of the society, and there will also be a life-size reproduction of a typical amateur station of the period 1911-13. A modern amateur station will be in operation at the exhibition and the call-sign will be GB3ERD.

The exhibition will be opened at 3 p.m. on 3rd April, by the Mayor of Derby, Alderman Miss M. E. Grimwood-Taylor, whose father, S. Grimwood-Taylor, was one of the founders of the society.

Being the oldest amateur radio society in the U.K. (now with 245 paid-up members) it has many historic pieces of equipment and components and an effort is being made to set up a permanent museum.

### 100MHz m.o.s. i.c.s

Metal oxide silicon transistors have been successfully made at the Hirst Research Centre (G.E.C.) with channel lengths of only  $1\mu\text{m}$  against the 6 to  $10\mu\text{m}$  normally employed in commercial m.o.s.

integrated circuits. Apart from increasing packing density by a factor of ten it is estimated that 100MHz integrated circuits can be made with these devices and could mean that m.o.s. will soon be knocking t.t.l.

The short channel transistors were made from masks produced by Cambridge Scientific Instruments Ltd using electron beam methods. The ion implanted source and drain regions have a low feedback capacitance giving the device its high operating speed.

The work had been carried out under a Ministry of Defence contract and is part of a much larger programme of research into m.o.s. devices being carried out at the Hirst Research Centre.

### Top of the short-wave pops

The International Short Wave Club reported a record response to their poll, held once every three years, to establish the most popular short-wave station. Votes have been received from most countries in the world including China and the U.S.S.R. Each voter could give five-votes for his first choice, four for his second and so on. It was found that many individuals did not go any further than a first choice. Of the 30,836 valid votes cast (4,712 arrived too late) Radio Australia

won with 7,010 votes followed by the B.B.C. World Service with 4,493. Third and fourth were Radio Nederland (3,877) and Voice of America (3,711), followed by Deutsche Welle (1,269), Radio Japan (1,051) and Radio Canada (1,018).

### Printed circuit patent

A decision taken recently in the House of Lords ended a ten-year legal battle by Technograph Printed Circuits (London). The decision confirmed that the patent taken out in 1943 by Technograph describing the manufacture of printed circuits was valid until it expired in 1967 after a seven-year extension. In this particular case the decision was against Mills and Rockley (Electronics) who had been sued for patent infringement by Technograph. However, the implications are much greater and writs have been issued against forty other British printed circuit manufacturers.

In America a similar ten-year law suit has been going on involving Technograph Printed Circuits' associate company Technograph Inc. although a decision has not yet been reached.

### Selective crystallization

Work is in progress at the Battelle Development Corporation, an international independent research institute, on a new process to produce m.o.s.t. microcircuits on a sapphire substrate. Instead of coating the sapphire with a film of silicon and using masking and etching processes to form active devices and interconnection patterns as is usual the new method does all this in one very complex process called selective crystallization.

The work so far has shown that the process is feasible but more research is needed before the method could be used for production.

*A new low-cost navigation system, installed on the Blue Funnel cargo liner Prometheus, intended to operate with the American navigational satellite system. The equipment, which was built by S.T.C., carries out dead-reckoning navigation between satellite passes.*



# Letters to the Editor

*The Editor does not necessarily endorse opinions expressed by his correspondents*

## Birthday greetings

Congratulations to *Wireless World* on completing its sixtieth year of publication. It was the world's first radio journal and has always been in the forefront in reporting technical developments in broadcasting. It is regarded as essential reading by most B.B.C. engineers.

The B.B.C. is a slightly younger organization, but during the forty-nine years of its existence, B.B.C. engineers have had the pleasure and privilege of contributing to *Wireless World* many articles on broadcasting and associated subjects. We value these close links with *Wireless World* and send it our best wishes for continuing success.

J. REDMOND,  
Director of Engineering,  
B.B.C.,  
London.

I am sure that thousands of radio and electronics engineers throughout the world will wish to congratulate *Wireless World* on reaching its sixtieth anniversary. Over this period, longer than any other journal in this field, *Wireless World* has accurately and honestly surveyed the broadest aspects of communications and electronics engineering, and initiated many designs of equipment using advanced techniques.

One of my most treasured possessions is a complete set of bound volumes of the journal. In these is recorded the whole history of the remarkable developments over the past sixty years, and perhaps the only regret is that the advertisements were not bound in for they record the changes in component design and the economics of the industry.

It was fortunate that my physics master at Westminster City School was a senior telegraphist in the R.N.V.R. during the first World War, and his enthusiasm for wireless and line communications continued into the school classroom and laboratory. Soon he had his students winding hundreds of feet of insulated copper wire on 4in diameter paraffin impregnated cardboard tubes and sending us off to Lisle Street, which was already the mecca for radio components. So in place of visiting the school tuckshop and buying

Fives balls we spent our pennies with Kate Raymond who sold ebonite end plates, aluminium vanes, 2BA rod and brass spacers, with which we made variable condensers. Farther along the road was Will Day, an outstanding inventor of cinematograph apparatus, who saw the potential development of the radio industry, and who later financially assisted John Logie Baird with his early television experiments. Will Day was the source of galena crystal, and, if one had the money, quite sophisticated micrometer adjustment cat's whisker crystal detectors. Then a trek to Shoreditch to buy some continental headphones from Ted Rosen who later became the chairman of Ultra Electronics, Ltd. Already the school had a good aerial consisting of two parallel 50 foot 7/22 s.w.g. copper wires on spreaders and a copper plate earth. Thus early in 1922 one was thrilled to hear the time signals from the transmitter on the Eiffel Tower.

These were the days of the amateur constructor and the founding of many radio clubs, the activities of which were regularly recorded in *Wireless World*. Although other journals sprang up to encourage the newcomer to wireless, already *W.W.* held the esteem of the relatively few professional engineers and the band of enthusiasts who followed the advanced constructional articles published in the journal. Many outstanding designs followed and one remembers the first to use a double ended screen-grid valve as an r.f. amplifier, and the using of Litzendraht inductances. Over many years *Wireless World* has sponsored designs for high-quality amplifiers of which The *W.W.* Push-Pull Quality Amplifier probably laid the foundation of the audio industry. Also in the early thirties members of the *W.W.* staff set up the first investigation into the performance of loudspeakers with a large baffle board mounted on the roof of the building and a microphone mounted on a mast.

Throughout the years since 1925 when I first made personal contact with members of the staff of *Wireless World*, I have always regarded it as a privilege to enjoy the friendship and advice of the successive Editors and their staff. Each has set a high standard and given devoted service to the journal, their readers and the industry. Many of the articles have been

contributed by world famous scientists, physicists and engineers and in spite of changing economic conditions and wars every issue has maintained a high and accurate standard.

Long may *Wireless World* continue to be published and its team of contributors and staff maintain the high standard of technical journalism that has made it the pre-eminent journal in the communications world.

J. C. G. GILBERT  
Northern Polytechnic,  
London N7 8DB.

## Stereo decoder using sampling

Referring to D. E. O'N. Waddington's article in the February issue, I have found that the second harmonic distortion level (0.6% at 30mV input) and the signal attenuation (9dB at 5kHz) can both be reduced by simple modifications to the circuit.

The de-emphasis network accounts for 5dB of the 9dB attenuation quoted, but what of the other four? As resistors  $R_{25}$  and  $R_{27}$  (Fig. 7 page 73) bias  $Tr_{10}$  and  $Tr_{11}$  near pinch-off, particularly for a transistor with low  $V_{(P)GS}$ , the operating point is one at which  $g_m$  is low and non-linearity comparatively high. Further, the low-value load resistor of  $2.2k\Omega$  implies low gain.

The 2V adjacent to the source of  $Tr_{10}$  suggests that transistors close to the lower limit of  $V_{(P)GS}$  were used in the prototype. The manufacturer's data sheet shows  $g_m$  to be 2mA/V, and hence a gain of about five times open-loop or 0.8 (-2dB) as a source follower. These figures were confirmed by measurement and a harmonic distortion figure of 0.1% obtained at 100mV r.m.s. level—relating to the source follower only.

Turning now to the f.e.t. switch  $Tr_8$  or  $Tr_9$ , it is important to bear in mind the inherent capacitance  $C_{gd}$  between gate and drain of the transistor. Were this fixed in value, its effect would simply be to cause an error between the input sample voltage and the output hold voltage across  $C_{12}$  or  $C_{17}$ . This error is caused by the transference of charge to  $C_{gd}$  on the trailing edge of the sampling pulse, and is minimized when  $C_{gd}$  is small and the hold capacitance as large as possible.

As the capacitance  $C_{gd}$  in practice is voltage-dependent in a highly non-linear manner the hold voltage error is also non-linearly related to the input sample voltage. This voltage is the instantaneous sum of the sample pulse amplitude and the composite signal at the source. The situation is summarized in the following expression

$$V_e = \frac{1}{C_{12}} \int_{V_1}^{V_2} C_{gd}(V) dV$$

where  $V_e$  is the error voltage, and  $V_1$  and  $V_2$  define the input amplitude limits.

Measurements on Mr Waddington's sample-and-hold circuit show an harmonic distortion figure of approximately 0.5%

to 100mV r.m.s. output across  $C_{12}$ , and an attenuation of 2dB was obtained.

To minimize both attenuation and distortion, the hold capacitor can be increased in value so long as it is able to become fully charged during the sampling interval. This is to say  $5CR = 250ns$ . No value is given for  $R_{ds(on)}$  in the BFW10 data but it seems this is below  $300\Omega$ , so that  $C_{12}$  and  $C_{17}$  can be increased to 180pF. Taking this opportunity to consider  $R_{ds(off)}$ , it is essential that the sampling pulse at the gate has an amplitude of at least 8V to ensure that limit transistors switch off as required, and this necessitates an increased supply voltage. Taking all these factors into consideration the following modifications can be made.

1. Increase supply to 12V and feed the drains of  $Tr_{10}$  and  $Tr_{11}$  from a 20 to 24-V supply.
2. Omit  $C_{15}$ ,  $C_{18}$ ,  $R_{24}$  and  $R_{26}$ , directly coupling the drains of  $Tr_8$  and  $Tr_9$  to the gates of  $Tr_{10}$  and  $Tr_{11}$ .
3. Increase  $C_{12}$  and  $C_{17}$  to 180pF.  $R_{25}$  and  $R_{27}$  to  $22k\Omega$  and  $R_{28}$  and  $R_{29}$  to say  $33k\Omega$  while reducing  $C_{19}$  and  $C_{20}$  to 1.5nF. These modifications reduce susceptibility to spread in f.e.t. characteristics, reduce distortion to 0.05% and reduce attenuation in the sample-and-hold and source follower sections to less than 1dB.

It is desirable to disable the sampling action for monophonic reception, by open-circuiting  $R_{13}$  and  $R_{21}$  for example, otherwise noise components around 38kHz are translated to audible frequencies causing appreciable degradation of signal-to-noise ratio, especially in view of the triangular noise spectrum associated with an f.m. system.

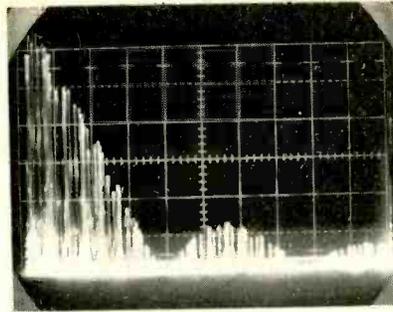
This type of decoder is very attractive in terms of cross-talk performance, suppression of 19kHz and 38kHz components, and reduction of subcarrier sidebands. Perhaps fuller advantage of the technique could be taken by using a phase-locked-loop sampling oscillator, which would enable both setting adjustments and wound components to be eliminated.

Incidentally, one of the transistors suggested for  $Tr_8$  to  $Tr_{11}$ —the BFW10—has a  $V_{(P)GS}$  max of 8V—dangerously close to the supply of 9V.

D. R. BIRT,  
Oxted,  
Surrey.

In his letter in the March issue D. E. O'N. Waddington states that for a sample-and-hold network it "... is not the 'sampling' but the 'hold' which causes noise harmonics to be heterodyned into the audio bandwidth" and that "The mark-to-space ratio of the sampling waveform has very little to do with the interference introduced. . ."

It must be remembered that sampling is a process whereby each component in one signal is multiplied by each component in the other signal, producing sum and



difference frequencies. It follows that the difference frequency produced when h.f. noise is sampled at a high rate will be l.f. noise.

Further, because the h.f. spectrum content of the sampling waveform is a function of mark-space ratio, the amount of l.f. noise produced will depend on the mark-space ratio of the sampling signal. The hold network does not appreciably alter the frequency content in the audio bandwidth as it does not perform a non-linear function. Above the audio bandwidth the h.f. signals are attenuated by the hold function.

The photograph shows the approximate response of Mr Waddington's sample-and-hold circuit on a scale of 0-100kHz. Zero amplitude occurs at integral multiples of 38kHz. The response has a  $|\sin x/x|$  shape and it is interesting to note that at 15kHz the theoretical response is 2.4dB down, agreeing very closely with the practical result.

I agree that low-pass filtering the input signal will prevent the generation of audio noise. The low-pass filter should ideally have a linear phase response to maintain high-frequency separation. I think Mr Waddington's suggested 80kHz is a little high—the 2nd harmonic of 38kHz will be able to operate on noise down to 61kHz to produce audible noise.

Improving the h.f. response of the decoder to reduce the 2.4-dB loss at 15kHz may be difficult if suppression of h.f. signals is to be maintained.

R. T. PORTUS,  
Rolls-Royce Ltd.

## Components for constructors

I must add my heartfelt agreement to the comment on this topic contained in "World of Amateur Radio" (*W.W.* Jan. 1971). Only in recent years have I been able to even contemplate building many pieces of equipment that captured my imagination when a teenager in the 1950s. Alas now that I can afford such luxuries I am faced with the near impossibility of finding parts.

My own immediate problem is quite simply that I wish to build a receiver covering from, say, 10 to 2,000m in the usual bands. I would like the receiver to have a reasonable sensitivity but not be in the communication receiver class. Basically, I feel able to "design" such a receiver along conventional lines.

Tuning coils can be obtained, and older readers will be delighted to know that Home Radio still include the famous

Wearite "P" coils in their catalogue. Alas no one to my knowledge markets a suitable dial or tuning scale covering the above ranges to match a specified tuning capacitor gang and the Wearite coils or for that matter anyone else's coils.

There are some of us who wish to build such equipment from scratch and who do not want to buy a kit complete to the last length of wire.

All I want to do is to build a reasonably simple receiver and my mind boggles at the trials in store for anyone brave enough to consider the *Wireless World* colour receiver. Construction will be a simple task compared with obtaining replies to enquiries about component availability. It is evident that people like myself must be a dying race and it is for this reason that supplies seem to have disappeared from the market. Some of us might even be prepared to pay through the nose for what we require if only someone would realize we exist and set out to exploit us!

F. BRIAN KYLE,  
Workington,  
Cumberland.

About a year ago we formed a buying group with a number of other dealers\* in the London area which we provisionally call "Group One". Its primary object is to buy components at the best prices in reasonable quantities. There are several secondary aims such as exchange of surplus stock and exchange of information. To some extent this action has been forced on us because we wished to buy certain items that wholesalers do not wish to handle and the manufacturers will only sell in quantities that are beyond the pocket of one dealer to buy. But I would like to stress the fact that this is not aimed at distributors or wholesalers (I for one, have always believed that they do a useful job and earn their money); in fact any small wholesaler or equipment manufacturer would be welcome to join. I feel sure that you will agree that this is a desirable scheme as ultimately it means the Group can offer your readers a greater range of goods at the lowest prices. Initially we were going to limit it to about 20 dealers (not on account of any closed shop principle, but because we thought, quite wrongly, that we could not handle the administration of a larger number). Now we would like to offer membership to any bona fide trader in the U.K. and I would be very grateful if you could make this generally known through the courtesy of your columns. At the moment there is no entrance fee or subscription. If anyone is interested please write to me at Home Radio (Components) Ltd, 234-235 London Road, Mitcham, Surrey CR4 3HD.  
ALAN SPROXTON,  
Mitcham,  
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\*Crescent Radio; EGE Components; Garland Radio; Hi-Fi & Components; H. C. Bros; Gurneys Radio; Newbury Radio; Odeon Smith; Lovers Radio Unlimited; Servio Radio; Electronics; Radio T. R. Radio; Watts Radio; Stan Reed; Tj Radio Comp

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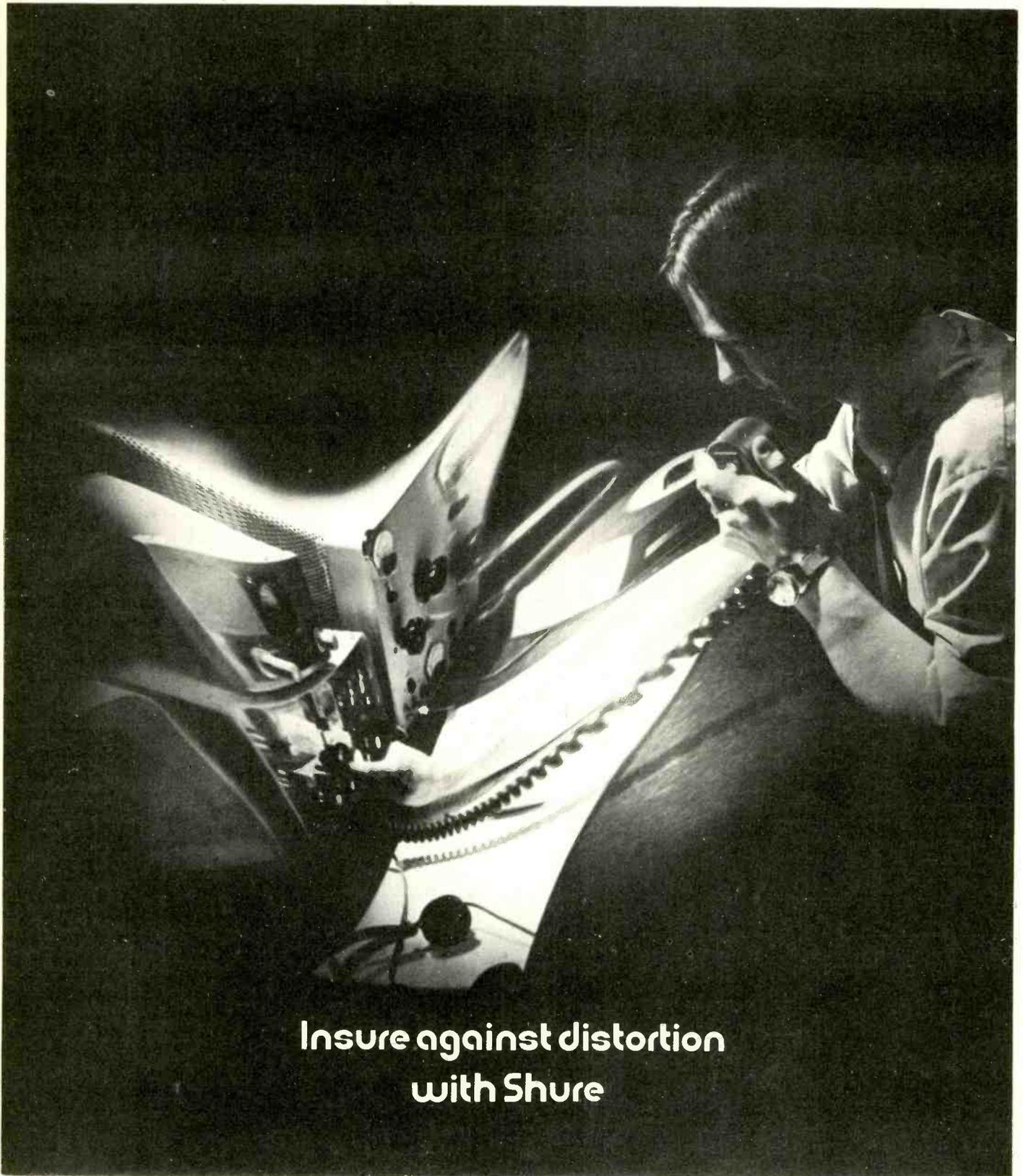
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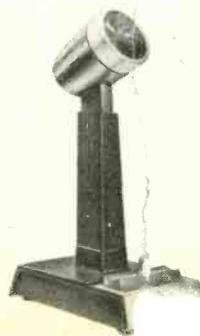
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# Karnaugh Map Display

**A low-cost instrument which may be used to assist in teaching logic and Boolean algebra**

by Brian Crank\*

Although this instrument employs the same basic principles as the earlier *Wireless World* Logic Display Aid (May to December 1969) any resemblance ends there. The present instrument is simpler and is very much cheaper. The circuit has been reduced to just four integrated circuits, three digital and one linear, and four transistors plus a few resistors, capacitors and diodes. The cost need not exceed a few pounds.

Mind you, the present instrument is not nearly so versatile as the earlier design although the display is more pleasing to the eye. The instrument will produce, on the screen of an oscilloscope, the Karnaugh map of any combinational logic circuit. If you are not completely familiar with Karnaugh maps a simple description will be found in the appendix to this article.

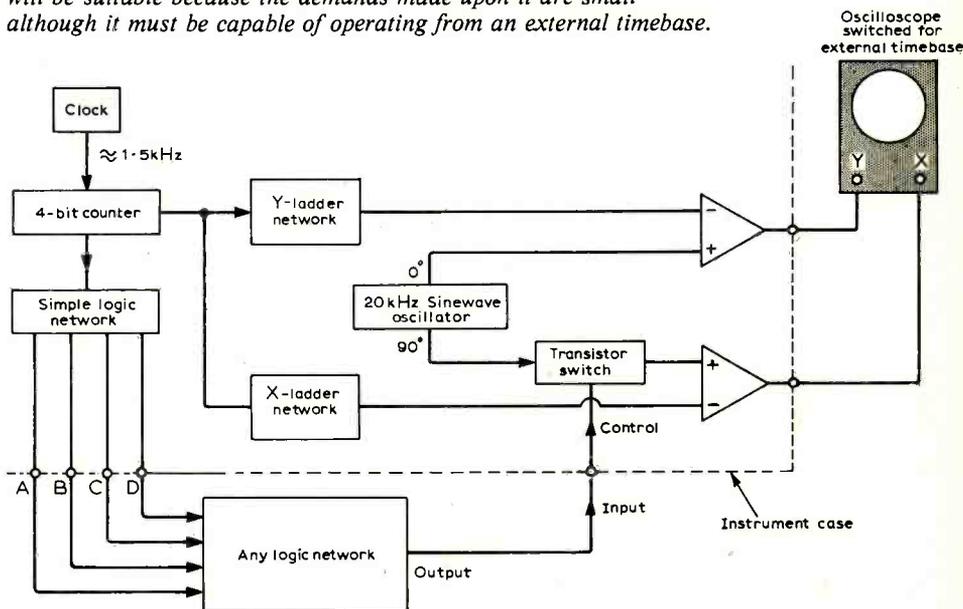
The reason for designing an instrument which will produce a Karnaugh map for any circuit is quite simple. The student is often taught Boolean algebra and logic through the use of the Karnaugh map. It completes the circle for the student to see a logic circuit producing the same map that was used to explain the operation of the circuit in the first place. In other words the theory and the practice can be brought closer together.

You may remember that in the earlier display the characters nought and one were displayed on the oscilloscope screen in the form of a pattern of dots. In the present design the characters are drawn as continuous lines exactly as you would draw them by hand. A typical display is shown in Fig. 11(a). Another advantage over the earlier design is that only two leads are needed between the instrument and the oscilloscope. These are the leads for *X* and *Y* deflection; an intensity modulation lead is not required.

As already mentioned the circuit is hybrid in that both linear and digital circuits are employed. Broadly the characters are positioned using a combination of both linear and digital techniques and the characters themselves are formed by linear circuits. The choice of whether to display a nought or a one at a particular position is taken by the logic circuit the Karnaugh map of which is to be displayed.

To use the instrument all one does is

Fig. 1. A simplified block diagram of the instrument. Practically any oscilloscope will be suitable because the demands made upon it are small although it must be capable of operating from an external timebase.



to connect it to the *X* and *Y* inputs of an oscilloscope, and connect any logic circuit to the instrument; the Karnaugh map for that circuit will then appear on the screen.

A block diagram of the instrument is given in Fig. 1. In brief, a clock pulse generator is used to drive a four-bit counter. The counter is split into two and each half drives a resistive ladder network. The ladder networks perform digital-to-analogue conversions and the resulting four-step staircase waveforms are fed to operational amplifiers which are used to drive the oscilloscope's *X* and *Y* deflection inputs. The oscilloscope is switched for external timebase operation. This produces sixteen dots on the screen arranged in a four-by-four matrix.

A sine wave oscillator, with a frequency much higher than the clock generator, produces two outputs which have a 90° phase difference. The sine wave corresponding to 0° is fed to the *Y* operational amplifier and the 90° waveform is fed to the *X* operational amplifier via an attenuator and a transistor switch. The result of the two sine waves on the screen of the oscilloscope is a vertical ellipse similar to the '0' printed here. The net result of both the sine and staircase waveforms is to dis-

play on the c.r.t. a four-by-four matrix of 0s. If the switch in the sine wave lead to the *X* operational amplifier is open there will be no horizontal sine wave component in the deflection waveform so on the screen will appear sixteen 1s. The 1 is formed by the sine wave input to the *Y* amplifier.

The counter that drives the ladder networks also drives a logic circuit which produces outputs that comply with the rules of a Karnaugh map. These outputs are used to drive the logic circuit you wish to display and the output of this logic circuit is used to control the 0/1 switch at the input to the *X* operational amplifier. Each section of the instrument will now be described in detail.

## Sine wave oscillator

The sine wave oscillator is used to produce a Lissajous figure which represents 0 in the display and to do this, as we have already seen, it must produce outputs at 0° and 90°. An early version of the instrument used a sine wave *RC* oscillator followed by a 90° phase-shift network. Although this worked it was unsatisfactory because it was necessary to specify close tolerance components for the frequency

\* Deputy Editor, *Wireless World*

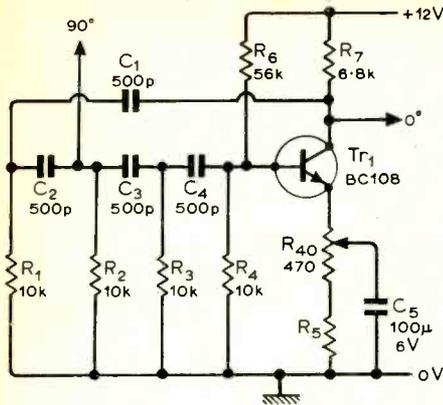


Fig. 2. The four-section phase-shift oscillator used to produce the characters which form the display. Operating frequency is about 22kHz.

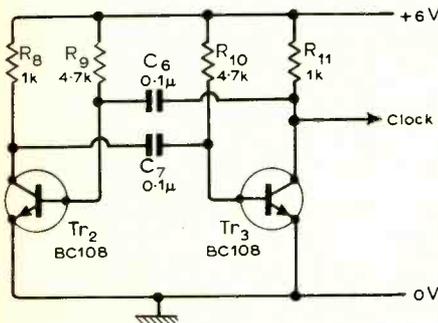


Fig. 3. Astable multivibrator clock generator which runs at about 1.4kHz.

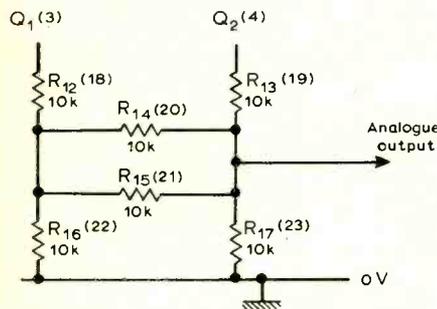


Fig. 4. The Y ladder network. Component reference numbers in brackets refer to the X ladder. The circuit converts the output of a counter into a staircase waveform by performing a digital-to-analogue conversion.

to be right for the phase shift required. An LC oscillator could have been used with the advantage that the frequency adjustment, to line the oscillator up with the phase-shift network, would have been no problem. However, coils, as well as being fairly bulky at the frequency we are interested in, are not the most popular items in constructional articles so it was decided to find a solution using RC circuitry.

The circuit employed is shown in Fig. 2. As can be seen it is a single transistor phase-shift oscillator. Normally a phase-shift oscillator employs three RC sections, each section phase shifting by 60°, to obtain the 180° phase shift necessary to obtain positive feedback and oscillation.

In the present design four RC sections

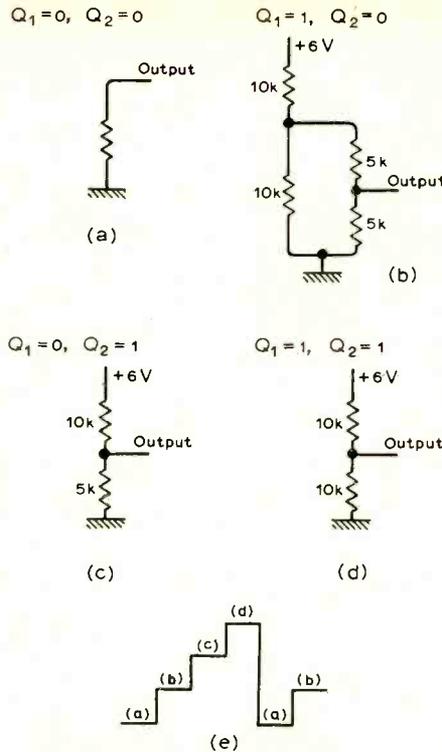


Fig. 5. The equivalent circuits of the ladder network for the four different conditions of the counter driving it.

are employed, each section shifting by 45° (4 × 45° = 180). It is now a simple matter to pick off the 90° signal after two 45° phase shifts at the output of the second RC section.

The potentiometer  $R_{40}$ , the only adjustment in the whole instrument, is used to vary the a.c. gain of  $Tr_1$  while maintaining d.c. conditions. The gain must just be enough to overcome the losses in the phase-shift network. If the gain is too low oscillation will not occur; if it is too high distortion will result. Potentiometer  $R_{40}$  is adjusted for a good sine wave output from  $Tr_1$ . The frequency of oscillation is about 22kHz but this is not at all critical.

**Clock generator and counter**

The clock generator is shown in Fig. 3. Little need be said about it as it is a conventional astable multivibrator which runs at about 1.4kHz.

The four-bit counter is formed by one t.t.l. (transistor-transistor logic) integrated circuit type SN7493N. This i.c. comes in the m.s.i. or medium scale integration class. It contains four J-K flip-flops and is connected as shown in the main circuit diagram (Fig. 10). The four flip-flops are cascaded to form a standard binary counter.

Looking at only the first two flip-flops, the outputs of which are called  $Q_1$  and  $Q_2$ , the following outputs are produced:

$Q_2$	$Q_1$
0	0
0	1
1	0
1	1

The outputs of the second pair of flip-

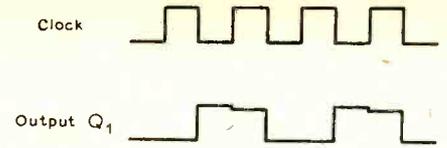


Fig. 6. Shows how the output of the t.t.l. binary counter is affected by the clock generator. The steps in the waveform are removed by a clamping circuit.

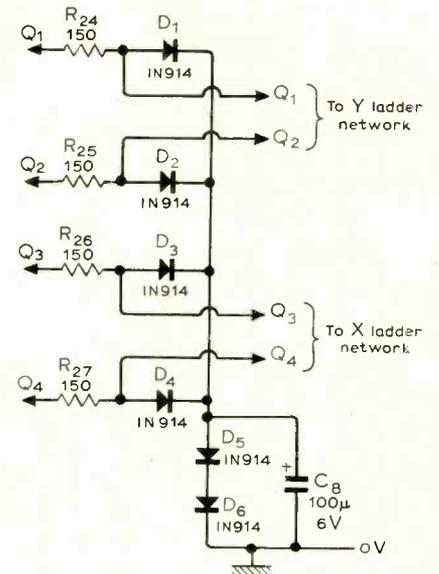


Fig. 7. The clamping circuit. The outputs to the ladder networks are the voltage drops across three forward-biased diodes in series.

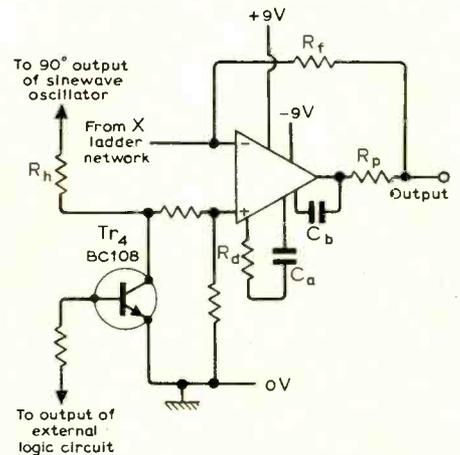


Fig. 8. The X deflection amplifier complete with the single-transistor I/O switch. The Y deflection amplifier circuit is the same but  $Tr_4$  and its associated components are omitted.

DO	0	1	1
CO	1	1	0
B			
A			
O	1		
1	1		
1	0		

Fig. 9. Karnaugh map edge coding. A graticule, the same as this drawing, should be made so that the display on the c.r.t. can be viewed through it.

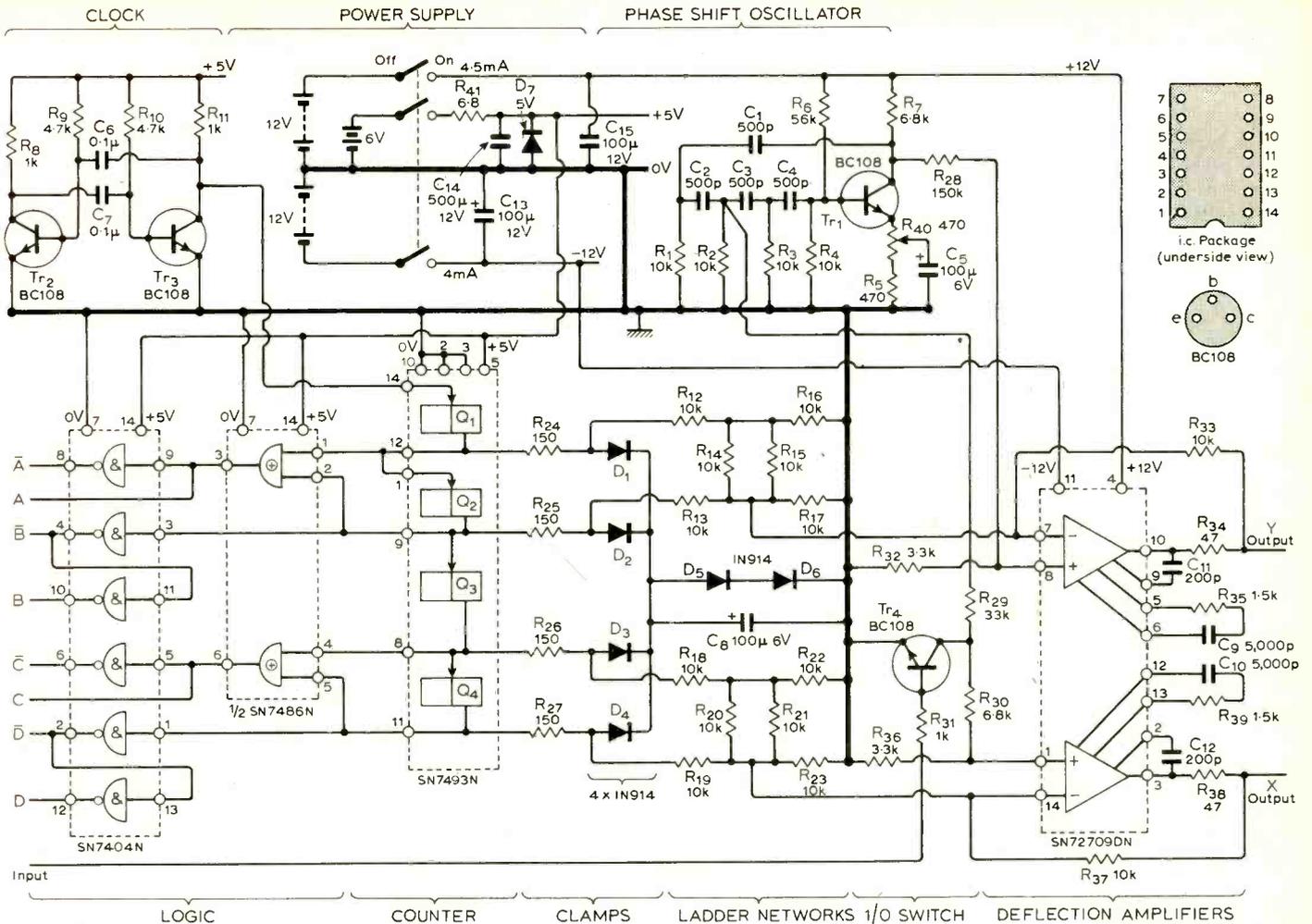


Fig. 10. The complete circuit of the Karnaugh map display instrument. The SN7486N actually contains four exclusive-OR gates, however, only two are used here.

flops,  $Q_3$  and  $Q_4$ , produce exactly the same output but at one quarter of the frequency.

**Ladder networks**

The two ladder networks are connected to the binary counter. When the flip-flop  $Q_1$  is at 0 the  $Q_1$  output is connected, via a saturated transistor, to the 0V line. When the output  $Q_1$  is at 1 it is connected via a saturated transistor to +6V.

The circuit of one ladder network is given in Fig. 4. The inputs  $Q_1$  and  $Q_2$  are switched according to the table given earlier. So Fig. 4 can be redrawn for each of the four states of the counter, so far as the output voltage is concerned (Fig. 5). If you would care to do the sums you will find that the output will rise from 0V in equal steps to produce a staircase.

**Clamping network**

Unfortunately the output of the flip-flops is not a good square wave. Although the rise and fall times are far more than adequate for the instrument the output of a particular flip-flop is affected by its input conditions. Fig. 6 illustrates this point.

The step in the waveform causes a corresponding step in the output of the ladder network which in turn causes certain characters on the display to appear double. The cure for this trouble is to add a clamp-

ing network which slices the top off the output from the flip-flops. This network is shown in Fig. 7.

The diodes  $D_1$  to 4 isolate the outputs of the flip-flops from each other and resistors  $R_{24}$  to 27 limit the current to a safe value. The output to the ladder network is now the voltage drop across three diodes in series.

**Operational amplifiers and 1/0 switch**

The well known operational amplifier type 709 is used in the instrument. The particular version employed (SN72709DN) is manufactured by Texas Instruments and includes two 709 amplifiers in a single dual-in-line package. The circuit of the X deflection amplifier is shown in Fig. 8. The Y deflection amplifier is identical except that the 1/0 switching transistor,  $Tr_4$ , and its associated components are omitted.

Resistors  $R_p$  and  $R_f$  combine to form the feedback resistor which sets the overall gain of the amplifier. Additionally  $R_p$  protects the amplifier from accidental short circuit of the output leads by limiting the output current.  $R_d$ ,  $C_a$  and  $C_b$  are frequency compensation components which ensure stability.

The BC108 ( $Tr_4$ ) is the switch which is controlled by the external logic circuit. It short-circuits the 90° output of the sine wave oscillator to ground when a 1 is

required on the c.r.t.  $R_h$  is of a sufficiently large value to prevent the switch from significantly affecting the oscillator itself.

**Logic circuit**

Imagine that the Karnaugh map of Fig. 9 is superimposed on the c.r.t. face. Because of the action of the previously discussed circuitry the c.r.t. spot first rests in the top left-hand square, it then moves to the next square down, then to the square below that until it reaches the bottom of the column. The spot then flies back to the top but this time to the second column. The process continues until all 16 squares have been scanned. The spot then goes back to the first square again and the process is repeated, such is the effect of the two staircase waveforms. Each square on the map corresponds to a particular state of the counter. For instance, the top left-hand square is scanned when the counter outputs are all 0, that is at the top of both staircase waveforms (both the X and Y amplifiers invert).

We also know that each square on a Karnaugh map corresponds to a particular set of variables as defined by the coding at the edge of the map (see appendix if necessary). We must ensure that when the spot is in a particular square that the set of variables represented by that square are available at the output of the instrument for

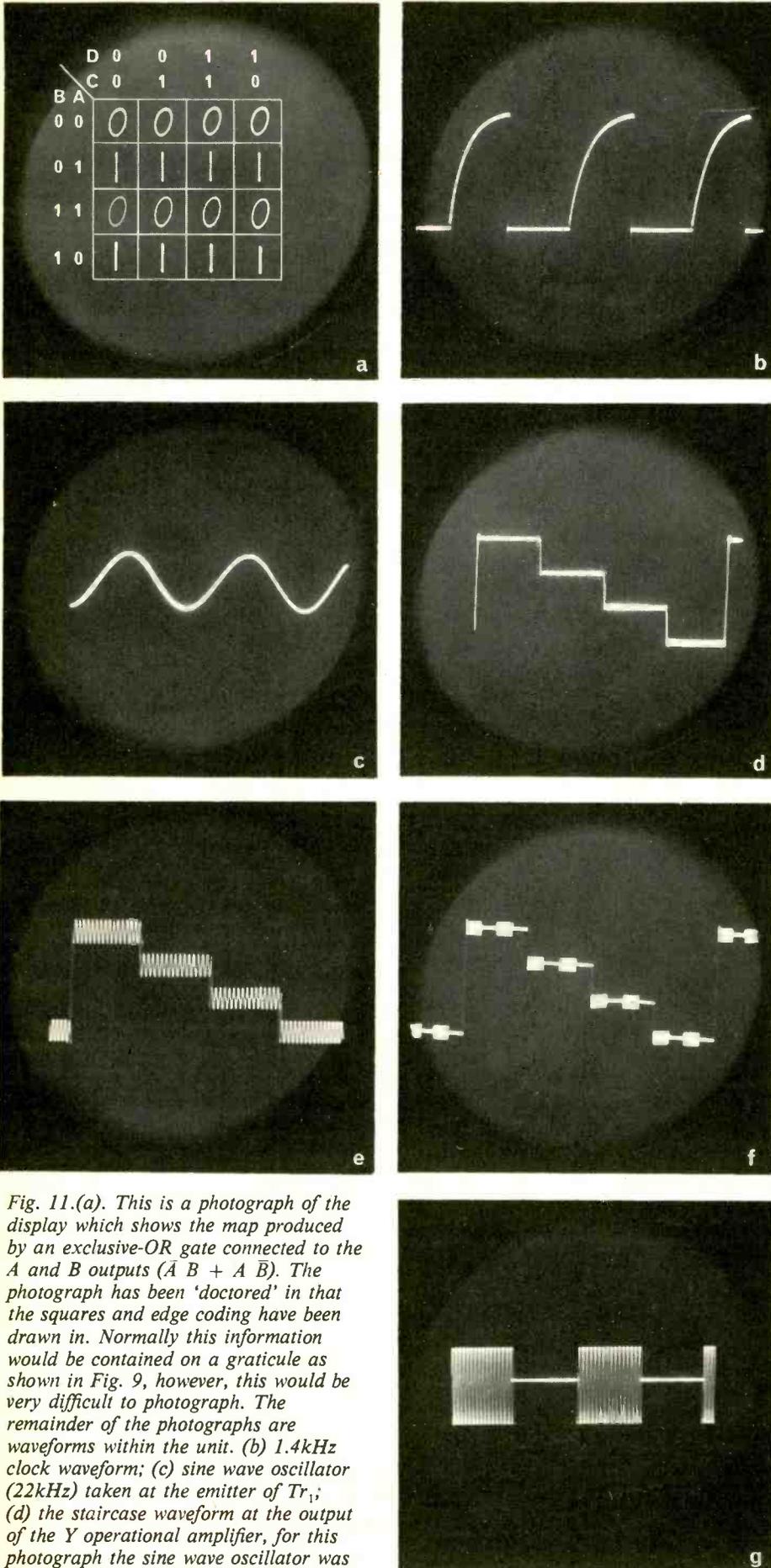


Fig. 11.(a). This is a photograph of the display which shows the map produced by an exclusive-OR gate connected to the A and B outputs ( $A B + A \bar{B}$ ). The photograph has been 'doctored' in that the squares and edge coding have been drawn in. Normally this information would be contained on a graticule as shown in Fig. 9, however, this would be very difficult to photograph. The remainder of the photographs are waveforms within the unit. (b) 1.4kHz clock waveform; (c) sine wave oscillator (22kHz) taken at the emitter of  $Tr_3$ ; (d) the staircase waveform at the output of the Y operational amplifier, for this photograph the sine wave oscillator was disabled; (e) Y deflection output when the display at (a) is being produced and (f) X deflection waveform under the same conditions; (g) waveform at the collector of  $Tr_4$  when the display at (a) is being produced.

feeding to the external logic circuit. We must therefore compare the output of the counter with the Karnaugh map edge codings and rectify any differences that occur.

Karnaugh map edge coding		counter outputs	
B	A	$Q_2$	$Q_1$
0	0	0	0
0	1	0	1
1	1	1	0
1	0	1	1

The above table compares the output of the Y counter with the map's A B edge coding. The last two terms are different and therefore some logic is necessary to correct this.

Firstly on examination we can say that  $Q_2 = B$  so a direct connection from the counter output  $Q_2$  will form the output variable B.

Also, on examination, it can be seen that:

$$A = Q_1 \cdot \bar{Q}_2 + \bar{Q}_1 \cdot Q_2$$

which is our old friend the exclusive-OR function. We have already stated that the X counter outputs,  $Q_3$  and  $Q_4$ , have the same outputs as  $Q_1$  and  $Q_2$  but at a slower rate and we can see that the Karnaugh map coding for C and D is the same as for A and B. We must therefore conclude that an identical logic function is required, namely

$$D = Q_4$$

$$\text{and } C = Q_3 \cdot \bar{Q}_4 + \bar{Q}_3 \cdot Q_4$$

The circuit of the logic section of the instrument can be seen on the lower left-hand side of the main circuit diagram, Fig. 10, and it can be seen that only two integrated circuits are required. The output variables, A, B etc., are buffered by simple inverters to prevent external connections from upsetting the operation of the counter. These inverters also provide the complement of the variables,  $\bar{A}$ ,  $\bar{B}$  etc.

### Complete circuit

Fig. 10 combines all the circuits discussed so far and therefore little need be said about it. The various waveforms present for a particular display are shown in Fig. 11. Because the sine wave oscillator and the clock are not synchronous flyback between characters takes a different route every time and is not visible on the screen at normal brightness levels. Because of this blanking (a Z connection to the oscilloscope) is not required.

### Construction

Making the unit is quite straightforward and no special precautions need be taken. A photograph of the layout employed in the prototype is given in Fig. 12; several components will not be found in this picture because they are mounted on the reverse side of the board.

It is important to connect pins two and three of the binary counter (SN7493N) to the 0V line. These pins are inputs to a gate which resets the counter. If this is not done the counter will be held at 0000 and the unit will not function. The only adjustment is  $R_{40}$  which must be set to give a nicely shaped 0. If you wish to adjust the size of the characters changing the value of  $R_{28}$

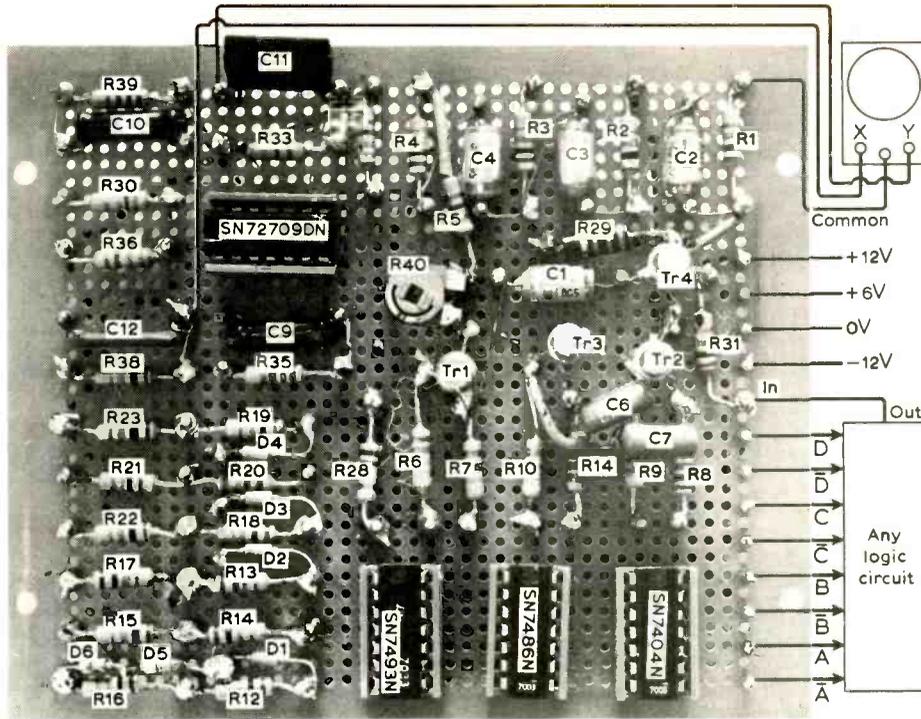


Fig. 12. A photograph of the prototype showing component positions. It should be noted that some parts have been mounted on the reverse side of the board and are therefore not marked. The integrated circuits are plugged into dual-in-line sockets which makes for easy removal.

will alter the height and  $R_{30}$  will alter the width.

**Appendix**

**Karnaugh maps:** The Karnaugh map is a means of pictorially showing all possible combinations of a number of two-state variables. Because of the way it is constructed it has other properties which make it possible to simplify Boolean expressions with the minimum of effort although it must be said that for more than four variables it is usually better to employ a more advanced method.

We will construct a Karnaugh map for four variables. The map will be the same as that displayed on a c.r.t. using the instrument described in the article. The basis of a Karnaugh map is a square. Each variable (usually labelled  $A, B, C$  and  $D$  for convenience) is allocated half the area of the square. To indicate the area occupied by a particular variable a simple edge coding system is employed. Fig. 13(a) shows the area occupied by the variable  $A$  and it is the area adjacent to the 1s under  $A$  in the edge coding. What is the area adjacent to the 0s under  $A$  in the edge coding? This is obviously the area representing  $\bar{A}$ . If the square of Fig. 13(a) is cut out and rolled into a cylinder the areas representing  $A$  and  $\bar{A}$  become continuous—but more about that later. In Fig. 13(b) the areas representing  $B$  and  $\bar{B}$  have been added. The square is now divided in four and each section represents one of the four possible combinations of  $A$  and  $B$ . From top to bottom, reading the edge coding, the sections are  $\bar{A}\bar{B}, A\bar{B}, AB, A\bar{B}$ .

You may have noticed that as you progress down the map, or up for that matter,

only one of the variables alters at a time and this still applies if the map is rolled into a cylinder again because  $A\bar{B}$  becomes adjacent to  $A\bar{B}$ .

In Figs. 13(c) and (d) the variables  $C$  and  $D$  have been added. If you consider only these two variables and roll the map into a cylinder the opposite way each section differs by only one variable. Reading round the tube so formed we get  $C\bar{D}, C\bar{D}, CD, \bar{C}D, \bar{C}D$  etc.

Looking at the map as a whole it is plain

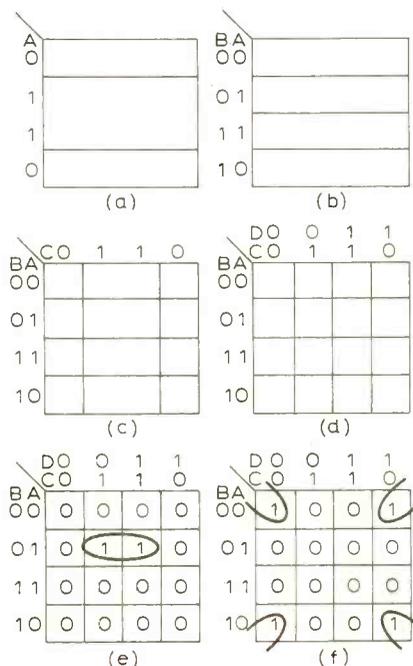


Fig. 13. The construction of a Karnaugh map and two examples. See text for full explanation.

to see that each one of the sixteen squares we have formed represents one of the possible combinations of the four variables. For instance the top left-hand square, as can be seen by the edge coding, represents  $\bar{A}\bar{B}\bar{C}\bar{D}$  and the bottom right-hand square represents  $ABCD$ .

But more important still is that adjacent squares, horizontally or vertically not diagonally, differ only in the negation of one of the variables. We have also proved, by rolling the map into a cylinder, that the top of the map is adjacent to the bottom and the left-hand-edge is adjacent to the right-hand-edge.

Two simple examples will show how these properties can be used to simplify Boolean expressions. Consider the expression  $\bar{A}\bar{B}C\bar{D} + A\bar{B}C\bar{D}$ . Draw a map as in Fig. 13(d) and put a 1 in the two squares representing the terms in the expression and an 0 in all the other squares. Because the 1s are adjacent to one another they are ringed as shown in Fig. 13(e). The simplified expression is derived by taking only variables which are common in adjacent terms. So  $\bar{A}\bar{B}C\bar{D} + A\bar{B}C\bar{D}$  reduces to  $\bar{A}\bar{B}C$ .

Fig. 13(f) shows the Karnaugh map for the expression  $\bar{A}\bar{B}\bar{C}\bar{D} + A\bar{B}\bar{C}D + A\bar{B}C\bar{D} + \bar{A}B\bar{C}D$ . All terms are adjacent and form a square of their own so only variables common to all four terms need be used. Therefore, from the map of Fig. 13(f):  $\bar{A}\bar{B}\bar{C}\bar{D} + A\bar{B}\bar{C}D + A\bar{B}C\bar{D} + \bar{A}B\bar{C}D = \bar{A}\bar{B}\bar{C}$

This brief explanation will serve to give the reader some idea of what a Karnaugh map is all about.

Next month a memory unit will be described which can be used with the Karnaugh map display unit, in place of the external logic circuit, to form an 'electronic blackboard'. Up to two Karnaugh maps can be stored, displayed or amended at will.

**Shopping List**

**Resistors**

All resistors, except the potentiometer, are 0.25W 5%.

- 10kΩ × 18
- 150 × 4
- 470Ω × 1
- 150k × 1
- 56kΩ × 1
- 33k × 1
- 6.8kΩ × 2
- 3.3k × 2
- 1kΩ × 3
- 47 × 2
- 4.7kΩ × 2
- 1.5k × 2
- 6.8Ω × 1

470Ω preset potentiometer.

**Capacitors**

- 500p × 4
- 5,000p × 2
- 100μ, 6V × 2
- 200p × 2
- 0.1μ × 2
- 100μ, 12V × 2
- 500μ, 12V × 1

**Semiconductors**

- SN7493N, 4-bit binary counter, × 1
- SN7486N, quad exclusive-OR gate, × ,,
- SN7404N, hex inverter, × ,,
- SN72709DN, dual op-amp, × ,,
- BC108 transistors, × 4
- 1N914 diodes, × 6
- 5V, 400mW zener diode × 1

**Miscellaneous**

- dual-in-line sockets, × 4
- Lektrokit board type LK141, × 1
- Lektrokit pins, × 100

# SONEX 71

## Hotel hi-fi show at London Airport

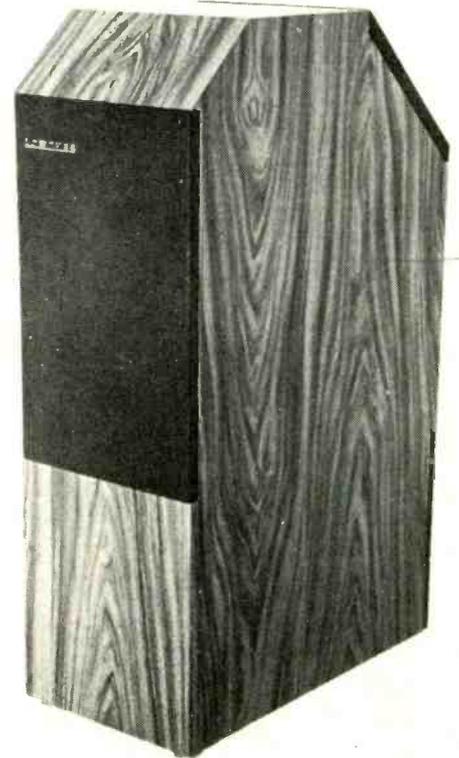
Over 150 rooms have been booked by more than 60 exhibitors for the second Sonex exhibition to be held at the Skyway Hotel, London Airport. The show will be open to the trade only on Wednesday 31st March and Thursday 1st April, between 11.00 and 18.00. It will be open to the public for the next three days, from 11.00 to 21.00 on Friday and Saturday and from 11.00 to 18.00 on Sunday April 4th.

Because of the postal strike admission tickets will not be required. Each visitor will receive a free sixteen-page show guide.

Features of the exhibition that are new this year include a 'Living with Hi-Fi' display arranged by *Homemaker* magazine, and the promise of considerable activity by Radio London.

### Brand names at the show

A.D.C.	Empire	Leak	Revox
A.K.G.	F.A.L.	Lowther	Richard Allan
Acos	Fane	Luxor	Rogers
Acoustic Research	Gabraphone	Metro-Sound	Rotel
Akai	Garrard	Mordaunt-Short	Sansui
Alpha-Arena	Goldring	Musitapes	Sheppard
Armstrong	Goodmans	National	Shure
Audio Packs	Grampian	Panasonic	Sinclair
Audio Technica	Harman-Kardon	Ortofon	Skandia
B.S.R. McDonald	I.M.F.	Peak Sound	Sonab
B & W Electronics	J.B.L.	Peerless	Sonotone
Bib	J.V.C. Nivico	Philips Electrical	Sugden, J.E.
Brenell	Jordan-Watts	Pickering	Telefunken
Cambridge Audio	K.E.F.	Pioneer	Teleton
Celestion	KMAL (Monks)	Poly-planar	Thorens
Connoisseur	Audio	Quad	Wharfedale
Decca	Koss	Radford	



*The Lowther Auditorium Acousta is an enclosure combining two sound sources—one forward, the other rearward. Using two such enclosures in place of a pair of conventional single perspective speakers gives increased solidity to stereophonic reproduction.*

Futuristic Aids will be demonstrating a loudspeaker from Fane Acoustics which employs a new ribbon unit for mid range and top frequencies and a 12in driver for bass.

Cambridge Audio have a new speaker called the Junior, and a 'slave unit' to boost the output of their amplifiers.

Rogers will be demonstrating a modified and improved version of their Studio Monitor loudspeaker which is being manufactured under licence from the B.B.C.



*An f.m. tuner from J. E. Sugden employs a varicap tuned front end and a four stage i.f. section with double tuned couplings. A switchable filter permits reduction of L-R information thus allowing separation to be traded for reduced noise on weak signals.*

# Elements of Linear Microcircuits

## 7: Radio- and intermediate-frequency amplifiers

by T.D. Towers,\* M.B.E., M.A.

We tend to think of silicon monolithic technology (in which complete circuits are produced in small crystal chips little larger than the dot on a printed 'i') as being very modern. Technologically this is so, but any reader who has access to Vol. 1 of *Wireless World* could well be surprised to find in the July 1913 issue a long article by Dr. A. E. H. Tutton entitled 'Crystals as Rectifiers and Detectors' examining crystal lattice structures in detail and pointing the way to the use of 'the very best procurable single individual crystals'. Some of the semiconductors examined by Tutton such as 'perikon', 'anastase' or 'brookite' may only evoke nostalgic memories from more elderly readers. On the other hand, unexpectedly, Tutton also devoted attention to pure silicon (so important now to monolithic integrated circuits), and to such sophisticated materials as silicon carbide and liquid crystals (important now in semiconductor opto-electronics).

The celebrated H. J. Round in a reply to Dr. Tutton in the August 1913 issue was curiously prophetic in his remark 'Crystals work well as wireless receivers'. In the 1970s we have indeed reached the stage where virtually the complete receiver circuitry can be produced in a silicon chip. The present article examines the state of the art for r.f. and i.f. amplifier microcircuits.

A survey of r.f./i.f. amplifier microcircuits in early 1971 showed about 100 commercial types available, all produced by semiconductor device manufacturers. Of these, virtually all are monolithic, although hybrid circuits are beginning to appear. Before we devote the remainder of the article to monolithics, it might be well to consider why hybrids are appearing alongside them.

Fig. 1 gives the circuit of the Newmarket NMC809A, a thick-film hybrid r.f./i.f. amplifier. The seeing engineer's eye will note that it is a d.c. coupled feedback pair  $Tr_1$ ,  $Tr_2$ , with an emitter follower buffer  $Tr_3$ , the transistors being 2N918 family devices with a typical frequency cut-off approaching 1,000 MHz. The circuit has a frequency response from d.c. to over 40 MHz. The resistor trimming possible with thick film

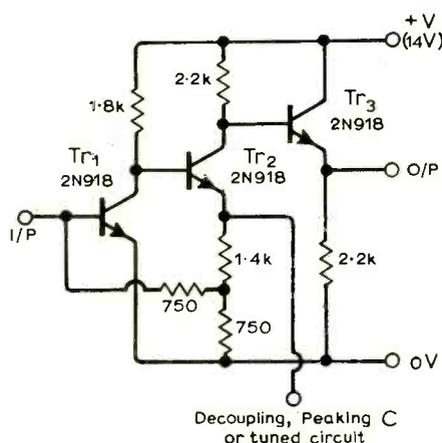


Fig. 1. Example of a thick film hybrid r.f./i.f. wideband amplifier; Newmarket Transistors NMC809A.

assembly gives it a very narrow gain spread; typically  $\pm 1$ dB in 22dB at 1 MHz. It can be used as a wideband amplifier or as a tuned r.f./i.f. amplifier up to 40 MHz. The separate connection to the emitter of  $Tr_2$  allows for full decoupling for low-frequency applications, or for a peaking capacitor for top-end video expansion, and series or parallel tuned circuits for band-pass or band-reject purposes. Designed to work on 14V with typically 14mA current, the NMC809A does not require any external biasing components, as bias levels are set up by resistor trimming during manufacture.

### Monolithic r.f. circuits

In conventional discrete component r.f./i.f. amplifiers it is normal to use a single transistor per stage, usually in a common-emitter arrangement, or, at very high frequencies, in a common-base arrangement. In both of these methods feedback from output to input can lead to instability and various neutralization techniques have to be used when high stage gain is called for.

Where higher stage power gain is wanted, some d.c. coupled arrangement of two transistors can be used instead of a single transistor between tuned circuits. Three configurations often met with are

the long-tail pair, the cascode arrangement and the d.c. feedback pair.

Now with discrete circuitry the single transistor common-emitter stage has such advantages in gain, noise figure and impedance matching convenience that you only come across the other arrangements mentioned where abnormal performance is required.

When we come to monolithic i.c.s, cost per transistor becomes less significant, being replaced by cost per stage. Here we find single transistors abandoned, and one

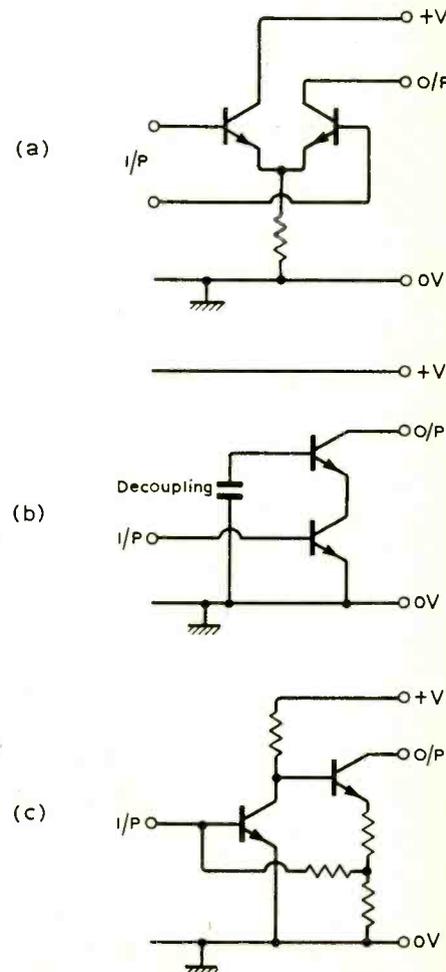


Fig. 2. Basic transistor configurations used in monolithic r.f./i.f. amplifiers; (a) long-tail pair; (b) cascode; (c) d.c. feedback pair.

of the three multiple-element arrays used. For ease of analysis, these are shown in basic form, without isolating and biasing networks, in Fig.2. Each has its advantages. The long-tail pair of Fig.2 (a) has fairly low noise, high power gain, high stability, simple biasing, inherently symmetrical operation, non-saturating limiting action, fast recovery from overdrive and easy interstage matching. The cascode of Fig.2(b) has low noise, high power gain, high stability, and easy interstage matching. The d.c. feedback pair of Fig.2(c) is distinctive for low noise, large signal handling capacity and low power consumption.

**Long-tail pair**

The long-tail pair with a resistor tail is widely used in r.f./i.f. monoliths. One example is given in Fig.3, which is the circuit diagram of the Philips TAA380A. This is an i.f. amplifier suitable for use in TV intercarrier sound circuits and f.m.

broadcast receivers and comes packaged in the small multi-lead TO-5 ou/line metal can. Transistors  $Tr_1$  and  $Tr_2$  form a long-tail pair with tail resistor  $R_1$  and buffer isolating emitter follower  $Tr_3$  feeding to further long-tail pairs  $Tr_4, Tr_5$  and  $Tr_7, Tr_8$  to drive an output tuned circuit connected to terminal five. Transistors  $Tr_9$  and  $Tr_{10}$  provide stabilized voltages for close control of d.c. bias; reference voltages being provided from selected points in the series diodes  $D_{1-8}$  which are forward biased through  $R_{11}$ .

The resistive-tail long-tail pair is used as a basic element in many other commercial monolithic r.f./i.f. amplifiers such as the General Electric PA189, RCA CA3041, 3043, Fairchild  $\mu A717$ , Motorola MC1350, Siemens TBA120 and the Philips TAA570. Most of these are complex multi-stage circuits to provide a complete block of functions for a receiver.

Some monolithic multi-stage r.f./i.f.

amplifiers use a transistor instead of a resistor as the constant-current source for the long-tail pair. This three-transistor type of gain block can be seen forming repetitive elements in cascode in commercial microcircuits and Fig.4, the Philips TAA350 is an example of this technique. The long-tail pair with transistor tail,  $Tr_1, Tr_2, Tr_3$ , with its output buffer emitter followers  $Tr_4, Tr_5$  are repeated for the four stages. The bias for the tail transistor is provided from a common rail voltage defined by the forward voltage drop across the diode-connected  $Tr_{21}$  fed from the positive rail through a 2.1k  $\Omega$  resistor. The differential amplification with current-driven long-tail pairs gives high a.m. rejection making the amplifier suitable for use with very simple f.m. detectors.

Several monoliths have been made which can be used in a large number of different ways. Fig. 5 gives the circuits of a number of the more common commercially available ones. These can, in most cases, be used as a cascode or a transistor-tail long-tail pair.

All the circuits of Fig.5 are basically a balanced emitter coupled transistor pair with a third transistor providing the emitter current for each of the pair. Such a configuration would be costly to fabricate with discrete components because of the difficulty of getting adequately matched balanced pairs and the need for separate biasing networks stable with temperature and supply voltage variations. With monolithic fabrication these difficulties do not arise. The different examples in Fig.5 reflect different manufacturers' design approach to versatile microcircuits with many circuit connection options.

The RCA CA3053/3028 of Fig.5(a) uses resistor networks to bias the tail transistor  $Tr_3$ . The Motorola MC1550 of Fig.5(b) features a diode in the biasing network for temperature stability; the Fairchild  $\mu A703$  of Fig.5(c) uses two biasing diodes; and the Amelco 911 (also National Semiconductors devices LM171/271/371 of Fig.5(d) uses as many as three bias semiconductors for maximum stability. Fig.5(e), the Signetics NE511, is a different approach and offers two amplifier stages in the one package together with one biasing diode. This gives the equipment designer great flexibility for special circuit requirements. Another example of extreme versatility is the RCA3004/3020 shown in Fig.5(f).

**Cascode circuits**

So far we have not mentioned cascode operation. It will be found that in monolithic microcircuits a true two-transistor cascode arrangement is almost never found. This is because a cascode circuit can be made up by taking a transistor-tailed long-tail pair such as Fig.5(a) and using the tail transistor  $Tr_3$  as the common emitter input of the cascode pair and one of the balanced pair as the common-base output. Some of the circuits of Fig.5 have direct access to the base of the tail transistor and can be used as

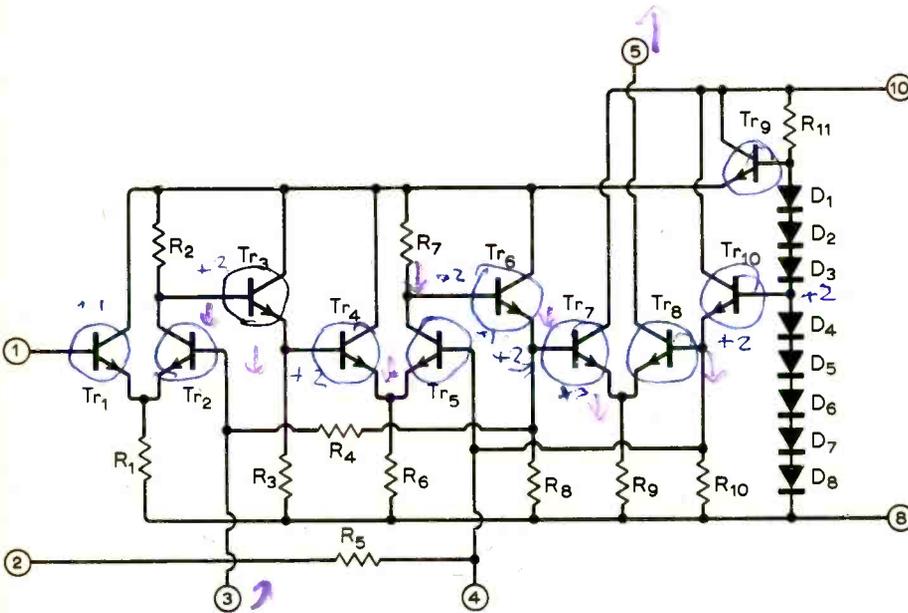


Fig. 3. Commercial i.f. amplifier microcircuit illustrative of use of resistor-tail long tail pair as basic amplifying element in monolithic microcircuit construction (Philips TAA380A).

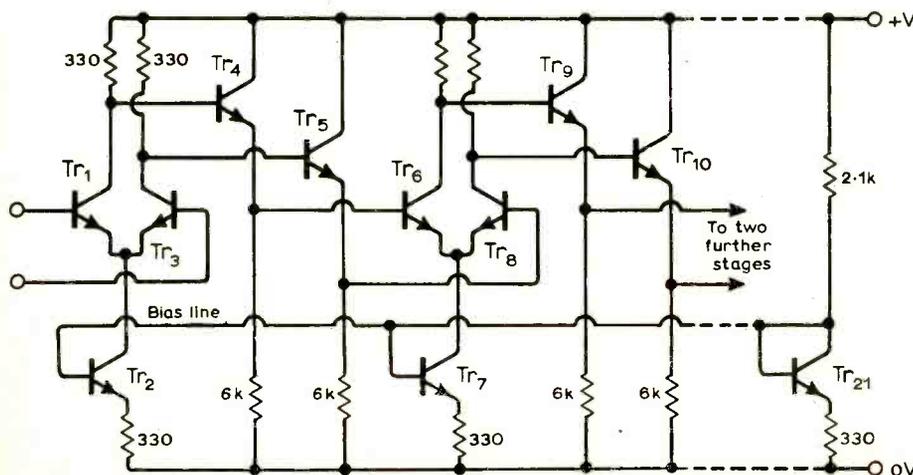


Fig. 4. Monolithic r.f./i.f. amplifier long-tail pair with transistor constant current tail (Philips TAA350).

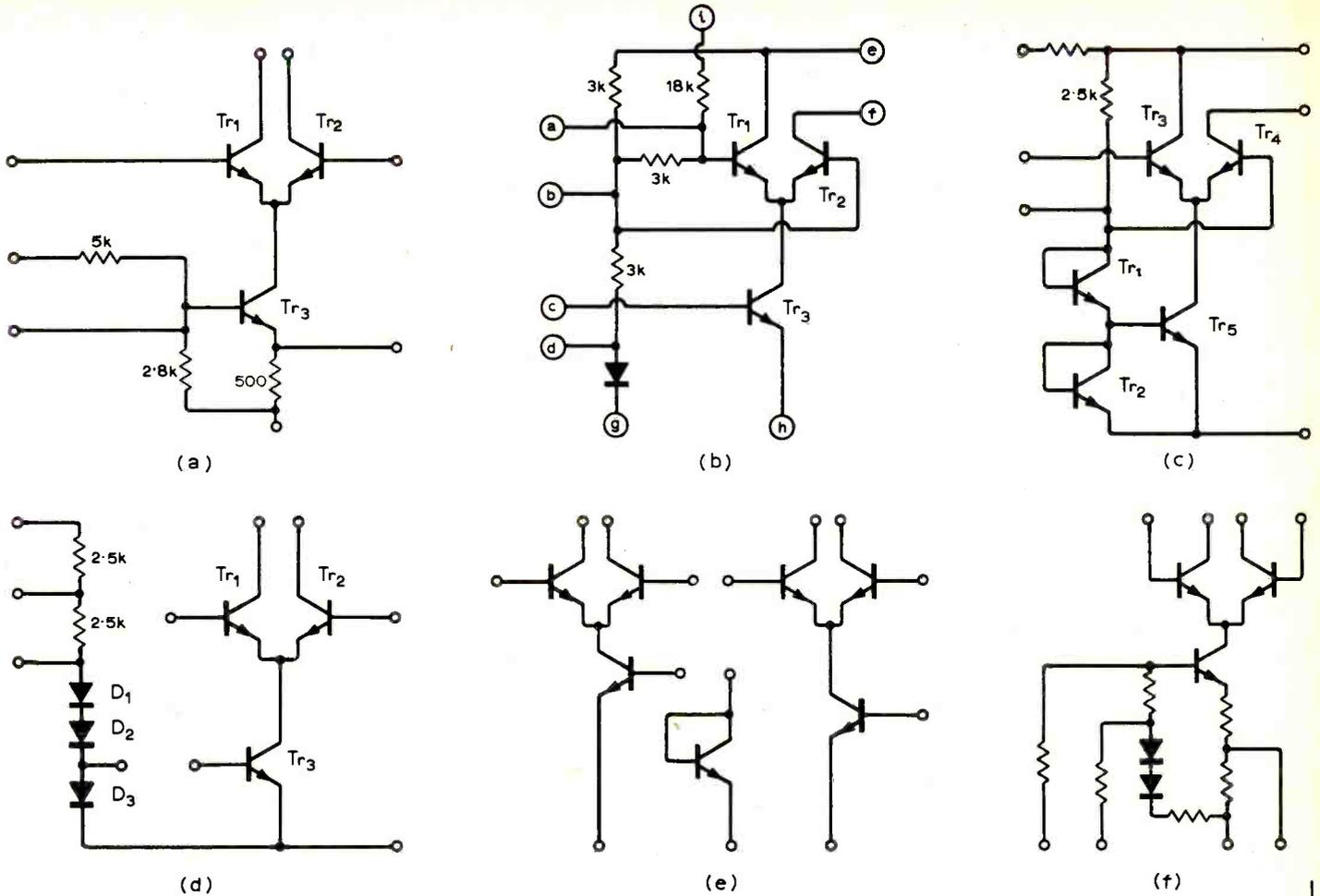


Fig. 5. Commercial single-stage monolithic r.f./i.f. amplifiers using long-tail pairs with transistor tails; (a) RCA CA3053/3028; (b) Motorola MC1550; (c) Fairchild  $\mu$ A703; (d) Amelco 911; (e) Signetics 511; (f) RCA 3004/3020.

effectively in cascode as in long-tail pair mode (where the input is applied to one of the balanced pair). One feature not to be overlooked is that the three-transistor configuration (whether it be cascode or long-tail pair) always leaves an unused terminal which can be employed to apply automatic gain control.

Perhaps the easiest way to understand the versatility of the transistor triplet is to look at one specific example, say the Motorola MC1550 whose basic circuit is given in Fig.5(b). This microcircuit can be connected, for example, as a cascode 60MHz tuned amplifier in the arrangement of Fig.6(a) or as a long-tail pair 10.7MHz i.f. amplifier as in Fig.6(b).

**D.C. feedback pair**

The cascode and long-tail pair are not the only configurations used by monolithic manufacturers. The d.c. coupled feedback pair shown for discrete circuitry at Fig.2(c) above has advantages that have led to its adoption by some manufacturers. One example of this is the Plessey SL612 r.f. amplifier (Fig.7). The design is essentially an emitter coupled d.c. feedback input pair  $Tr_1, Tr_3$  providing d.c. bias from the emitter of  $Tr_3$  to the input of  $Tr_1$  via a  $5k \Omega$  resistor. The d.c. coupled pair is followed by an emitter follower  $Tr_5$  providing feedback into the emitter of  $Tr_1$  via a  $525 \Omega$  resistor. The overall circuit gain is 20 dB Low noise is ensured by running  $Tr_1$  at low current and good signal handling by the overall feedback. Effective a.g.c. control of 50dB is achieved via  $Tr_8$  and

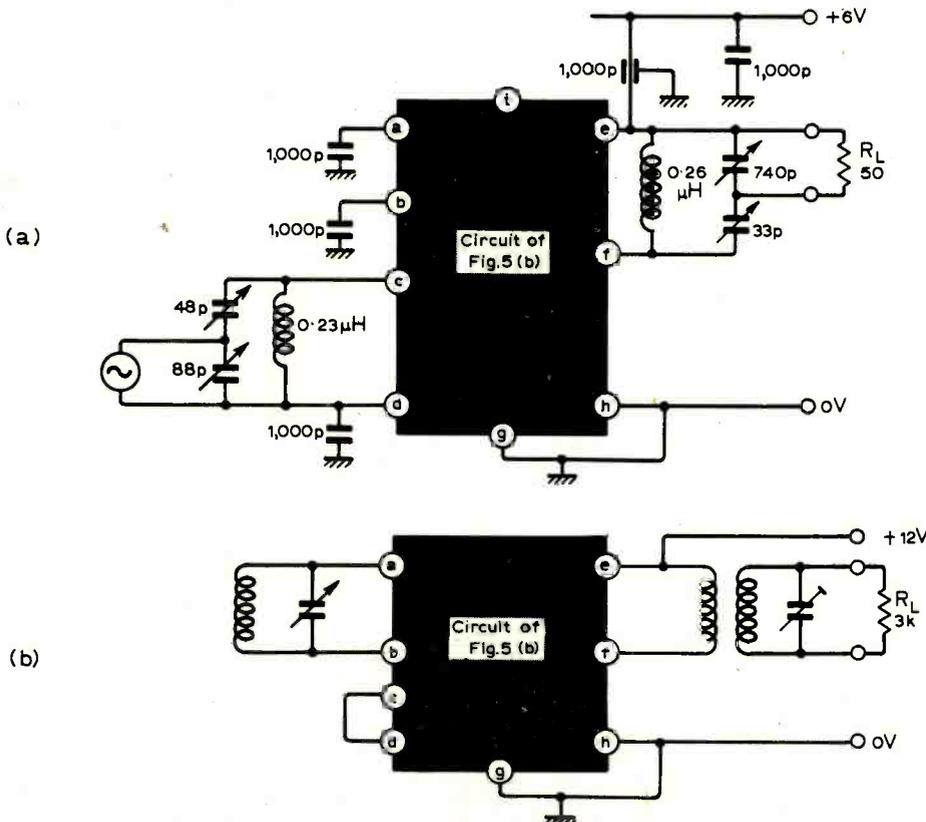


Fig. 6. Alternative cascode and long-tail pair tuned amplifier arrangements of Motorola MC1550 microcircuit; (a) cascode 60MHz, 30dB 0.5MHz bandwidth; (b) long-tail pair 10.7MHz f.m. i.f.

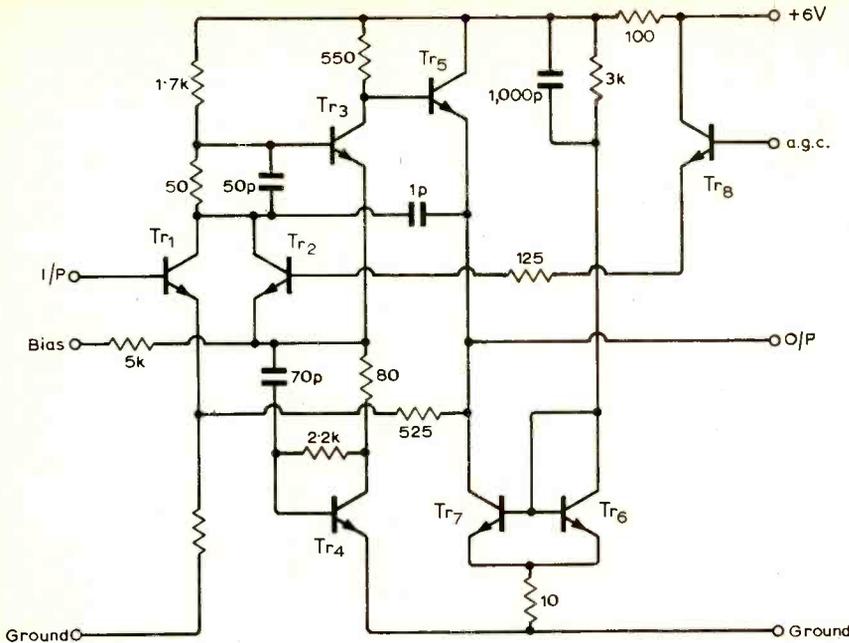


Fig. 7. Commercial r.f. amplifier monolithic microcircuit using basic d.c. coupled feedback pair configuration: Plessey SL612 r.f. (2-76MHz) tuned amplifier.

Table 1  
Directory of r.f./i.f. microcircuit manufacturers

Company	Microcircuit type numbers
Amelco (Teledyne)	911
Fairchild	$\mu$ A703/717/719
General Electric (USA)	PA189
Intermetall	TAA710
Marconi-Elliott Microelectronics	M316
Mitsubishi	M5142P
Motorola	MC1100/1350/1352/1550
Mullard (Philips)	TAA350/380/380A/450/570/640
National Semiconductors	LM171/172/271/371/372/703
Newmarket Transistors	NMC 809A
Plessey	SL501/502/503/551/552/553/610/611/612
R.C.A.	CA3002/3004/3005/3006/3028/3041/3042/3043/3044/3053
S.G.S.	L103, TAA661/730
Siemens	TAA981/991, TBA120/400
Signetics	NE510/511, SE510/511

$Tr_2$  and power consumption is some 15mA on a 6V supply. The circuit is optimized to give a 150MHz cut-off frequency ensuring satisfactory operation over the 2-76 MHz communications band.

To help readers see what is commercially available in the way of r.f./i.f. amplifier microcircuits, Table 1 lists the major manufacturers whose products of this type are on the U.K. market, with a selected list of types known to the author.

(to be continued next month)

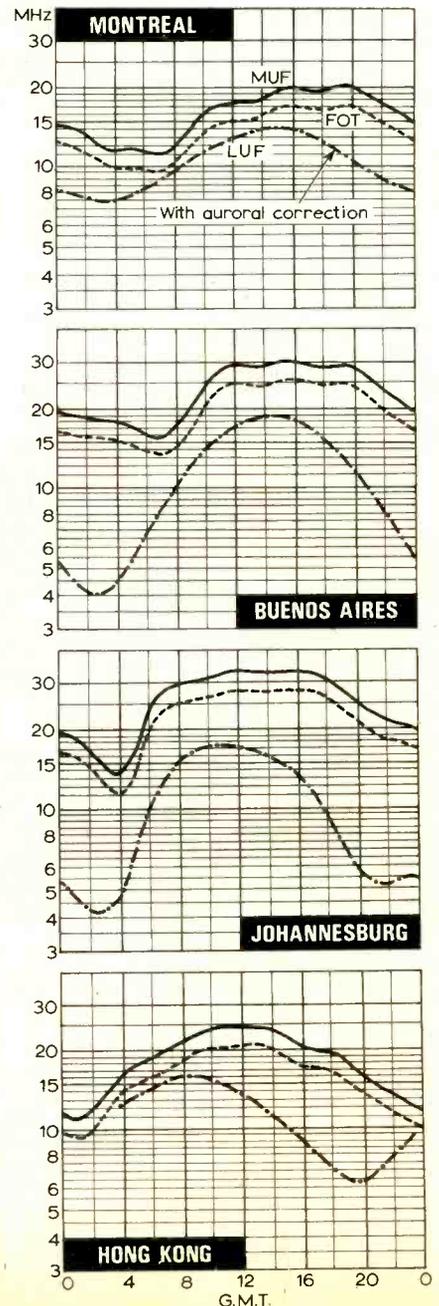
The wide range of microcircuits available to the equipment designer vastly simplifies the production of tuned amplifiers from 100kHz to 100MHz, whether for fixed frequency i.f. use, with high gain, low current consumption and narrow bandwidth, or for variable r.f. use, with low noise, large signal handling, a.g.c., without substantial change of characteristics, and without a performance change across the tuning band.

## H.F. Predictions—April

The charts are based on an Ionospheric Index of 96, the corresponding sunspot number being 83. Similar predictions were made for April 1970 but the observed index value jumped to 130 and remained high the following month. There was a similar trend in 1969.

Transequatorial paths have their highest MUFs during equinox months and conditions should be good above 20 MHz. Evening fading is relatively independent of season and cycle on the South African path but is worse during this season on others. Poor conditions are expected from the Far East midnight to 09.00 G.M.T. and North America will be liable to several days of weak signals 06.00 to 16.00 G.M.T.

MUFs apply to both directions of a route while LUFs are for reception in the U.K. only.



# Don't Look Now

## Sampled Data Controlled Systems, and . . . .

by Thomas Roddam

Some years ago, when the book *Principles of Feedback Design* was in preparation, I did not dream of discussing a topic which, I then thought, was way out of our world. The airlines tell me the world is getting smaller, but if it is, there is still the simple fact that it gets more packed every day. Having started on non-linear systems, and found that the describing function is relatively easy to use, we must examine the sampled-data systems because we need this method for practical equipment. We need it, in fact, for power supplies.

Simple sampled-data systems we use in everyday life. You come home, look at the thermometer, or just guess, and having sampled the room temperature you either switch on a heater, or you don't. There are two sampling modes which follow. Either you take a data sample, am I hot or cold, at regular intervals in the natural breaks of the telly programme, or you operate with a relay dead-band characteristic: it's getting too hot in here, and off goes the heater. It's so tempting to continue with this example until it collapses in confusion. There is one feature which can be examined. Suppose that you have a very powerful heater, and that it is a very cold day. With regular sampling you may never be comfortable: it gets too hot before the next sample is due, if the heater is on; if it is off it gets too cold. This is the characteristic behaviour of a non-linear loop with an intolerable instability.

The theoretical study of sampled data systems is complicated by a number of factors. The sampling may not be exactly regular in period: this is a very common situation. The sample may not be taken, as the simple theory demands, in an infinitely short time. I do not propose to be picky about this, because this is merely an introduction to the subject. Furthermore, I have some numbers in mind. A particular system contains, in the loop, an inductor which is nominally 50 mH at 8 A d.c. An inductor of this kind, of course, is really 50-70 mH at best, and may be up to 200 mH at lower values of polarizing current. And then there is a big electrolytic capacitor. Details of the sampling data process are trivial compared with the large range of uncertainty of the response of the linear system. It does not make engineering sense to impose close tolerances on these components just to enable the theory to be made more precise.

In most of the literature it would seem

that the forward path and the feedback path are shown as combined. This is because so much of the theory has been carried out in terms of control systems. For regulated power supplies this means, quite simply, that we connect the monitoring voltmeter in a different place. The essential bit of a linear regulator can take the form shown in Fig. 1. There are refinements, but they do not affect what we are considering. Normally we regard the voltage across the output terminals as the object of study, but in fact it is the voltage at point B which is the significant term so far as the regulator is concerned. The long-tail pair can be considered as a combiner and the complete unit rearranged into the form of Fig. 2. The actual output is then given by  $V_B$ /(pot' ratio). It is just the same for an audio amplifier, although we may put frequency dependent terms into the divider path. Stability depends on  $\mu\beta$ , not on either  $\mu$  or  $\beta$  alone.

A sampled data system begins by looking at the error signal. This is done by the sampler. The output from the whole system will be a continuous one, and so the system must have a continuous instruction supplied to it. This is provided by a hold circuit, which stores the latest news of the error, provided by the sampler, until the next sample is taken. You plan your Sunday by the sea on the basis of Saturday night's weather forecast: you can even get it in stored form, in the evening paper. The overall system takes the form shown in Fig. 3. In this the box marked "operate" would be marked ( $\mu\beta$ ) if we left out the sample and hold processes, and we should study the term  $(1 - \mu\beta)$ , the difference signal, the error signal.

The elegant way of studying the stability of a system characterized by a function  $F(s)$ , or  $f(t)$ , is to study the roots of the equation

$$F(s) = 0.$$

In a simple world, all we know is  $f(t)$ , the way the system behaves when we give it a momentary impulse, but we find that we are urged towards the use of the Laplace transform and then to the Cauchy theorem, so that we finish up with a result which is simply the Nyquist diagram. It is usually fairly easy to find  $F(s)$  for a linear system in the world of electronics. If we can find a form of  $F(s)$  for the sample and hold part

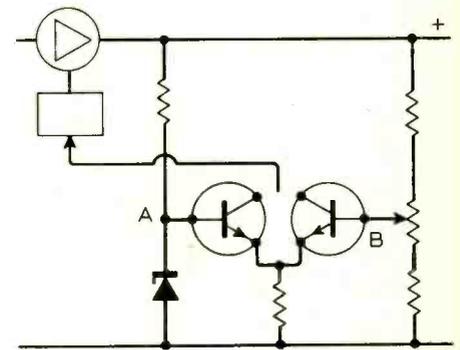


Fig. 1 Heart of a power supply regulator.

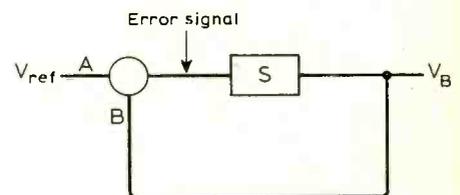


Fig. 2. Symbolic view of Fig. 1.

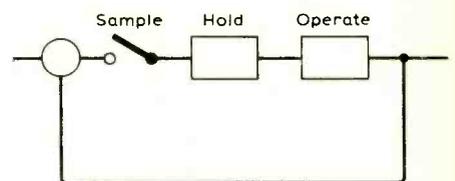


Fig. 3. Sampled data control system.

of Fig. 3 we shall be able to proceed absolutely along the established lines. One way of attacking this would be by the route of the describing function. The alternative is a bold frontal attack to determine the behaviour in time of the hold output when a standard signal is applied to the sampler. If the Laplace transform can be used on this process we have just two functions,  $F_1(s)$  and  $F_2(s)$ , for the two parts of the circuit which are in tandem, and overall we have

$$F(s) = F_1(s) \cdot F_2(s).$$

As it turns out, the Laplace transform which is obtained for a sampling system is an infinite series. I suppose this could have been expected, because the sampling process is one which implies infinite currents to reset the holder in an infinitely short time. A modified transformation, which is called

the Z-transformation and which is related to the Laplace transform, is used. It has a rather interesting effect on the stability rules.

We had better look at the process we want to put into mathematical terms. Fig. 4 shows how an input signal is first converted into a train of pulses and is then processed by the hold circuit, a circuit which is often called a box-car circuit. If we think of the conversion of 4(a) into 4(d) in terms of the describing function there are two points which are obviously fairly true. The first point is that

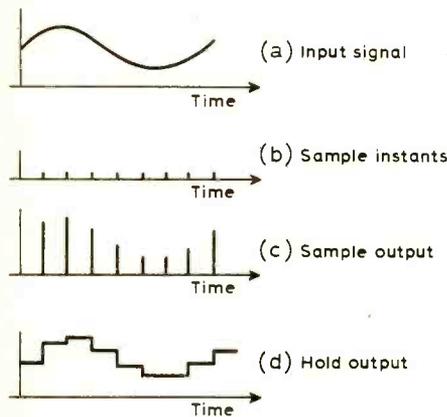


Fig. 4. Sample and hold process.

the describing function is not a function of amplitude, as it is with dead band or saturation non-linearities. The sample-and-hold operation is a linear one, in this sense. The second point is that the sample-and-hold operation provides a delay of about one-half the sampling period. This delay does not depend on the input frequency so that it is not a phase angle. Thus far, then, the s.h.o. behaves very much like a transmission line for the purpose of its describing function. Things get rather odd, however, if the input frequency is increased to around one-half the sampling frequency. If the samples are taken at 0 deg and 180 deg the output will be zero: if they are taken at 90 deg and 270 deg the s.h.o. will show a small gain. The output will show characteristic slow beats. This can be interpreted by treating the whole thing as a modulator, with input frequencies of  $f_i$  and  $(f_s + 2f_s + 3f_s \dots)$ , the spectrum of the sampler. The term  $2f_s - f_i$  is very close to  $f_i$  when  $f_s \approx f_i$ , and this is one way of looking at the beat source: another is to fix attention simply on  $f_s - f_i$ . From the first, however, we see that we are likely to get this sort of thing happening when  $f_i = 2f_s$ , and so on right up the spectrum.

Two consequences can be guessed. The first is that any attempt to find a simple describing function in the form  $G(s)$  will need to deal with an infinity of odd spots. This is because our sampler is what we might call an infinity generating function. The second is that we may get trouble with those beats in a practical system. Hold circuits do not normally hold for ever: if we think of a power supply filter as a hold circuit it will do its job nicely at 100 Hz, getting rid of ripple, but if there is an instability near the sampling frequency which produces an input, will the beat frequency get suppressed? The describing function technique is at its

safest if one is throwing away harmonics. Here there is a nasty danger of subharmonics, and in practical systems it is found that oscillations at a subharmonic of the sampling frequency are, indeed, produced.

The formal approach is, as we have said, by means of the Z-transform. This is developed in the following way. Suppose that the input to the sampler is a waveform  $f(t)$ . The output is a series of pulses, separated by time  $T$ . The  $n$ th pulse, at time  $t = nT$ , is quite simply  $f(nT)$  and is fully expressed by

$$f(nT)\delta(t - nT)$$

Where  $\delta(t)$  is the impulse function, a unit impulse at time  $t = 0$ , or, for  $\delta(t - nT)$ , at time  $(t - nT) = 0$ , or  $t = nT$ . As we know from our delving into the Laplace transform, the impulse function is  $\mathcal{L}$ -transformable. In fact we can write

$$\mathcal{L} f(nT)\delta(t - nT) = f(nT) \exp(-nTs)$$

This takes advantage of the fact that while we are looking at this pulse,  $nT$  is a constant, and so  $f(nT)$  is also a constant.

The whole pulse series which comes out of the sampler is the sum of all the individual pulses:

$$\sum_{n=0}^{\infty} f(nT) \exp(-nTs)$$

Now we write  $z = \exp(Ts)$  and the pulse becomes:

$$\sum_{n=0}^{\infty} f(nT)z^{-n}$$

We call this  $F^*(z)$ , the Z-transform of  $f(t)$ , and write

$$F^*(z) = \mathcal{Z} f(t)$$

Just as when we are dealing with functions of the complex frequency,  $s$ , we take the ratio of input  $I(s)$ , and output  $O(s)$ , and call its inverse the transfer function

$$\frac{O(s)}{I(s)} = G(s)$$

so in sampled systems we can take  $F_{in}^*(z)$  and  $F_{out}^*(z)$  and write

$$\frac{F_{out}^*(z)}{F_{in}^*(z)} = G^*(z)$$

the sampled transfer function. It is a special kind of system gain. Fortunately it is related to the ordinary meaning of gain. The mathematics is of a kind best studied in privacy by consenting adults, with long expressions all over the place. The final answer is elegant:

$$G^*(z) = \mathcal{Z} \mathcal{L}^{-1} G(s)$$

Thus all you do is work out  $G(s)$ , with all those  $(s+a)$  and  $(s^2 + as + b)$  terms that come from  $X, R$  and  $L, C, R$  circuits, find the inverse Laplace transform, using tables if you are idle, and then find the Z-transform, for which you should have another set of tables.

Although it is not uncommon for the hold function to be provided by a network of capacitance and inductance there are practical as well as theoretical circuits in which the box-car hold system is used. It is a circuit element of a rather special kind. It has what we may call a frequency response, that is to

say a transfer characteristic expressed in terms of  $s$ , which is obtained by using the Laplace transform on the behaviour in the time domain. It is rather unusual, in circuit theory, to find ourselves working this way round. Again the mathematics is a mass of long expressions, but in the end we obtain a transfer function for the hold circuit,  $G_h(s)$ , given by

$$G_h(s) = (1 - \exp(-sT))/s$$

Remembering that  $s$  is a generalization from  $j\omega$ , we see that we have a sort of  $j\omega T$ , and the exponential function gives us a power series of this, a harmonic series of the sampling frequency.

Although I do not expect that, from this article alone, readers will be in a position to design sampled data systems, it is nevertheless of some value to know the form which some basic functions take when they are Z-transformed. The obvious one to begin with is the unit step, the most basic function which can be meaningful. Although we took the unit impulse for our normal kind of circuit, sampling a unit impulse is an exercise in probability which I will not attempt, especially as we would be considering the chance of two infinitely short pulses coinciding. The Z-transform of the unit step, however, is quite easy. All the pulses are the same height, and

$f(nT)$  is a string of 1, 1, 1, ...

This makes

$$\begin{aligned} F^*(z) &= \sum_{n=0}^{\infty} z^{-n} \\ &= 1 + 1/z + 1/z^2 \dots \\ &= z/(z-1) \end{aligned}$$

It is more elaborate than the Laplace transform, which is simply  $1/s$ .

A key waveform is the one we get from a single root on the real axis, for which the  $s$ -function is  $1/(s-a)$ . For this the inverse Laplace transformation gives us

$$f(t) = \exp(at)$$

Each pulse in the train which comes out of the sampler is  $\exp(aT)$  the size in the preceding one:  $a$  is usually negative, of course. The Z-transformation gives us

$$\begin{aligned} \mathcal{Z} \exp(at) \cdot u(t) &= \sum z^{-n} \exp(anT) \\ &= \sum (z^{-1} \exp(aT))^n \\ &= \frac{z}{z - \exp(aT)} \end{aligned}$$

Once all the analysis has been gone through, we have the overall sampled transfer function. In the familiar case of a linear system we, if we are using the Laplace approach, find  $\mu\beta(s)$ , and then substitute  $j\omega = s$ . Most of us go by a different route, to get  $\mu\beta$  as a function of  $\omega$  without using  $s$  at all. When we have this expression for  $\mu\beta$  we plot it out as  $\omega$  goes from 0 to  $\infty$ . We should in theory plot it from  $\omega = -\infty$  to  $\omega = \infty$ , but we only need to do this in some very tricky systems. Where is the critical point (1, 0), or if you don't slip the 180°, (-1, 0)? This is the Nyquist test for stability. In a previous article we saw that this trick of standing at the critical point and saying "are we surrounded?" is the equivalent of searching the whole half-plane for roots by

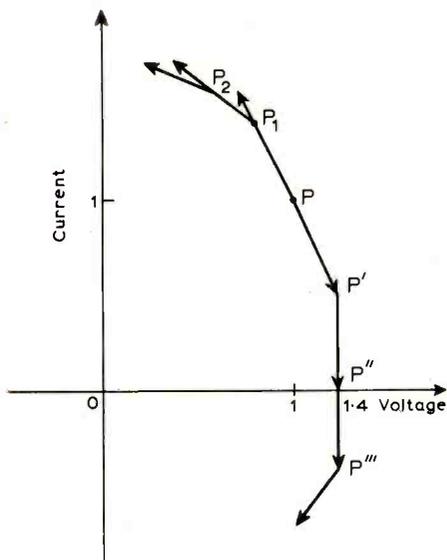


Fig. 5. Phase plane diagram.

going round it on a contour. We can be Indians or Wagon Train.

In a sampled data system we have

$$z = \exp sT.$$

Let us put  $s = j\omega$ , so that

$$z = \exp(j\omega T)$$

But  $\exp j\omega T = \cos \omega T + j \sin \omega T$

and as we vary  $\omega$  this means that  $z$  goes round and round on a circle of unit radius. Instead of varying  $\omega$  from 0 to  $\infty$  we need only go once round this circle. From then on, use Nyquist.

The underlying mathematics would seem to have evolved from the use of the Fourier and Laplace methods. Whenever one has a simple, powerful, mathematical technique one finds that there are mathematicians nibbling away at the foundations. Whenever the engineer says of a solution— that bit does not make sense, let's leave it out—the mathematician goes away to find a technique which will leave it out automatically. In the describing function method we leave out the harmonics, and sometimes land in trouble. The mathematics of the Z-transform comes from a study of where the Laplace transform goes wrong. There is less chance of applying the more rigorous theory which lies behind the Z-transform to functions which will have some odd feature which makes the procedure give the wrong answer. The difficulty, as always, is that the more general the rules, the harder it is to live with them: read the regulations at the park gate before you take the dog for a walk and you will see just what I mean.

Even if I had devoted ten times as much space to a very detailed explanation of how to use this classic sampled-data approach it would not have been a help in the real world. Practical sampled-data systems do not work with infinitely short sampling pulses, and the sampling pulses may not be evenly spaced.

One form of practical sampled-data system which is of both theoretical and practical interest is obtained when the sampling switch is closed for a finite time. This means that the sampler output is a normal pulse amplitude modulated signal, with pulses

which can pass reasonably well through at least a part of the linear system. The type of hold circuit is then probably not the zero order hold which produces the box-car effect, but a first or second order integrator. One important feature of this is that the theory must fit at the ends. When the pulse becomes very short the situation is the one we have just discussed, for which the Z-transform or other short-pulse techniques must give the same answer. If the pulse length is  $T$ , the sampler switch is closed all the time, and the answer must be the usual linear,  $\mu\beta$ , answer.

The sampler may produce pulse length modulation. I simply will not start on this. Except, there is one form of pulse length modulation which we must consider. Suppose that the sampler switch closes at a moment determined by the sign of the error signal, and is opened at regularly occurring times. This produces pulse length modulation with a modulated leading edge. Suppose also that the pulses are an odd shape. Who would want to build such a circuit? This is, in fact, just what one gets with thyristor regulation in phase-control and is not too different from some switching regulators.

Other switching regulators behave even more awkwardly. Both pulse frequency and pulse length vary over the controlled range. These, in fact, are probably best dealt with by using the describing function, treating them as relay circuits with hysteresis. But if you feed in some jitter signal deliberately they look like sampled data systems.

I want to look at some practical systems next month: there is not enough space here. There is, however, one other theoretical approach which I think deserves just about the space I have left if I am to stick to my standard length. The mechanical people call this the velocity-phase diagram technique, which does not mean much to us. It is chiefly used for rather simple systems, which are, in fact, just the sort of systems we want to use it for. That sentence expresses my meaning precisely: up with changes I will not put.

A typical circuit equation is

$$L \frac{dI}{dt} + RI + \frac{1}{C} Q = E$$

It is not necessary that  $L, C, R$  or  $E$  should be constant. We can always write:

$$\frac{dI}{dt} = \frac{E - RI - (1/C)Q}{L}$$

If  $I = dQ/dt$  we have

$$\frac{dI}{dQ} = \frac{dI/dt}{dQ/dt}$$

and thus 
$$\frac{dI}{dQ} = \frac{E - RI - Q/C}{LI}$$

This expression contains no mention of time, not even in the disguise of frequency, unless time is hidden in one of the parameters. The most likely place to find time is in  $E$ , but even the  $t = 0$  idea of the unit step can be eliminated in the plot which is adopted. Actually I never know what charge is: we used to catch it on little drops of oil. Volts I know, I've got a meter for finding them. And

$$Q = CV, \text{ so } \frac{dI}{dV} = \frac{E - RI - V}{(L/C)I}$$

The validity of one or two steps needs checking in each particular problem by running through this process writing not just  $C$  or  $L$  or  $R$  or  $E$ , but  $L_0 f(I)$  for an inductance which saturates, for example. The essential feature of this technique is that it is basically a practical one, working from the problem to the solution. The methods we normally adopt, like the Nyquist diagram, really start with the answer and we just test the problem against it.

The technique now is this. We choose our starting condition, when all the terms on the right-hand side are known. Let us begin by using the specific example we have, although this is not the only circuit we can handle in this way. To make life very easy, let us take  $E = 0, R = 1, (L/C) = 1$  and initial conditions  $V = 1, I = 1$ . We plot this point at  $P$  in a voltage-current diagram (Fig. 5).

Now at  $P$

$$\frac{dI}{dV} = \frac{-RI - V}{(L/C)I} = \frac{-I - V}{I} = -2.$$

We draw in an arrow to describe this slope. Let us go along this line to  $V = 0.8, I = 1.4$ , and call this  $P_1$ . Here

$$\frac{dI}{dV} = \frac{-1.4 - 0.8}{1.4} = -1.57$$

A new arrow is drawn to show the path from  $P_1$ . Using the approximation

$$\frac{dI}{dV} = -1.5$$

the next point is  $V = 0.6, I = 1.7, P_2$ .

Here 
$$\frac{dI}{dV} = \frac{-1.7 - 0.6}{1.7} = 1.35$$

There is no reason why we should not go on like this, building up a trajectory which shows how the circuit behaves. Except this: where does the arrow head come from? We must let time into the system just a little way. The time to go from  $P$  to  $P_1$  is

$$t_{PP_1} = \int_P^{P_1} dt = \int_P^{P_1} \frac{dV}{dV/dt} = \int_P^{P_1} \frac{dV}{I}$$

and as in this short stretch  $I \approx 1$  and  $dV = -0.2$ , the time in the movement from  $P$  to  $P_1$  is negative. We should head off in the other direction, instead of looking back into history. So we build up  $P', P'', P'''$  and in the end we get the spiral path sketched in Fig. 6, which must, of course, decay to the

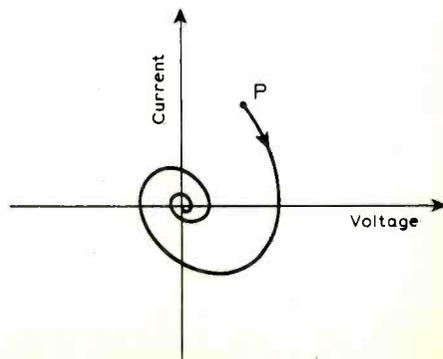


Fig. 6. The final picture.

origin, as we have no supply  $E$  to provide any final voltage. In this particular circuit a fixed voltage input would have given a spiral centred, if that is the word, on the point  $(E, 0)$ . This is obviously a stable system: in the long run it settles down. It is not a very exciting system, although it is easy to watch the spiral on an oscilloscope. If the brightness can be modulated it is possible to put time markers on as well.

In a non-linear system we can get up to all sorts of tricks. For simple folk like me a rather coarse stage by stage calculation is carried out. The example I am using now has been chosen so that it does not overlap the material I have in mind for the next article. Let us stick to our  $LCR$  circuit, but let us take some special cases. If we take  $R = 0$  we have

$$\frac{dI}{dV} = -\frac{V}{I}$$

$$IdI + VdV = 0$$

$$I^2 + V^2 = \text{const.}$$

This is the equation of a circle and I really do not think we need a figure. If we make  $R$  rather small we know that the oscillations die away very slowly. For convenience the system will be started off at the point  $I = I_0$ ,  $V = 0$ , and we shall get a spiral path which lies inside the circle. The first half-cycle is shown in Fig. 7 as the path from  $I_0$  to  $-I_1$ .

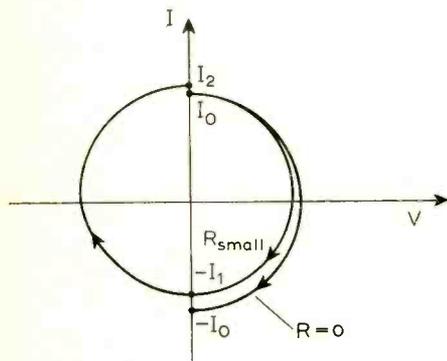


Fig. 7. Formation of a spiral.

Suppose that when  $V$  is negative the circuit is changed, to make  $R$  negative. The behaviour is then an expanding spiral, which I have drawn as  $-I_1$  to  $I_2$ . We are now back nearly where we started, and we can repeat the whole process. As the figure is drawn  $I_2$  is bigger than  $I_0$ , and so the spiral will get bigger and bigger. This, then, is a truly unstable situation.

In circuit terms this corresponds to an ideal class-B oscillator which drives for exactly half a cycle and decays for the other half. In terms of the  $s$ -plane the pole spends equal amounts of time on each side of the imaginary axis, but goes rather further into the right-hand half than it does into the left-hand half. The oscillator designer would disagree with the description "ideal". He wants the amplitude to stay constant. One way of doing this is to adjust the values of  $R+$  and  $R-$  very precisely, so that what we lose on the swings we make on the roundabouts. This is the function of a thermistor in such a circuit. There are two other ways, which are more in our line: the thermistor

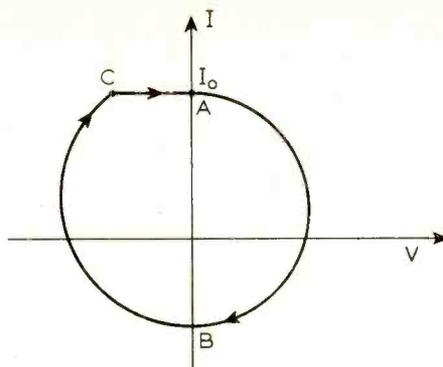


Fig. 8. Stabilization of oscillation by clipping at  $I_0$ .

is a linear operator, after all. We can use a limiter of some kind. It is a matter of taste how the limiting is done. A diode can be used to clip the voltage used in this analysis: a diode can also be used to clip the current. Examination of particular circuits will show how a non-linearity is added to each. Without giving any particular preference to any method, the sort of behaviour we get is shown in Fig. 8. This would be analysed in three pieces,  $A \rightarrow B$ ,  $B \rightarrow C$  and  $C \rightarrow A$ .

An alternative technique is to change the point at which  $R$  is switched from negative to positive. This is indicated in Fig. 9. From  $A$  round to  $B$  the damping is positive, and from  $B$  to  $A$  the spiral is a growing one. The change now takes place at  $-V_0$  and if this is chosen correctly the system remains in an equilibrium oscillation. It is, of course, a class-C oscillator and will be designed, with a d.c. loop study, to make  $-V_0$  self-adjusting. If the d.c. loop is not correctly designed the system has a different instability and is either a blocking oscillator or a squegger.

Circuits of this kind often have a dead-band characteristic. When this is so, the initial conditions become of great interest. There are two circles on the diagram of Fig. 10. If the circuit is set off at the point  $(I_1, 0)$ , inside the circle  $C_1$  the dead, or partly-dead, band means that the gain, in the describing function sense, is not enough to keep the circuit oscillating, and it just dies slowly. If the circuit is set off at  $(I_2, 0)$  it will begin to oscillate, and we have at first no restraint on the oscillation so that the spiral is a diverging one, until the circle  $C_2$  is reached, when the oscillation settles down. For the initial condition  $(I_3, 0)$  the limiting mechanism is rather over-working, bringing the

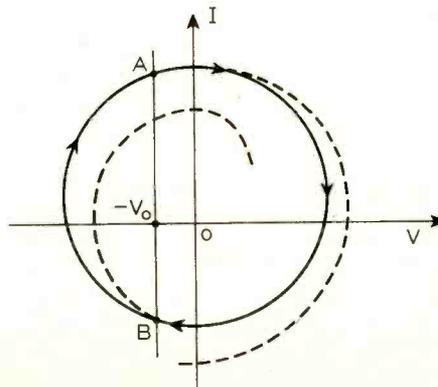


Fig. 9. Stabilization by change of transition voltage.

trajectory in towards  $C_2$ , the steady state oscillation.

For systems of reasonable simplicity these plots can be constructed on a point by point basis. They are also easy to produce by analogue modelling and to observe with an oscilloscope. Timing marks can be put along the trajectory, even when it is being constructed in this step-wise way. It is thus perfectly practicable to apply this method to a sampled-data system. The case against the method is one with which I have a good deal of sympathy. You are doing all this work on just this problem: you will have to do it all over again for the next problem. How nice to have general solutions. I am not satisfied, however, that the second, third, tenth problems of the same kind will involve the same amount of work. One great

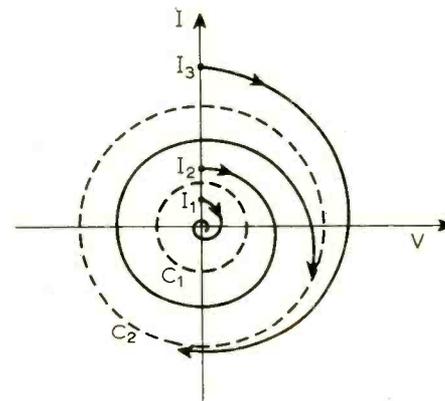


Fig. 10. Three initial conditions.

point in favour of this approach is that it keeps currents and voltages in your mind. They cannot slip off to infinity for an instant, leaving you with a sound analysis and a pile of dead transistors.

Non-linear circuit analysis, and especially that part of it connected with feedback systems, seems to be in a rather awkward phase of its existence. There exists a very large amount of theoretical work in the form of original papers, some of great subtlety. The textbooks, when they are up to date, deal with the more elegant refinements of linear theory before they come to non-linearity. Not only does this discourage the student, who, if he is studying in his own time, collapses before he reaches the non-linear material: the academic knows that non-linear systems are either trivial or too complicated to be subjects for an examination question. Meanwhile, we all need to build them. So let us stop thinking about Aspects of Non-linear Control Theory and start remembering, You, too, will go round the bend.

# Low Distortion Tone-control Circuit

## Bipolar transistors used in a Baxandall configuration

by P. M. Quilter\*

Now that very high quality transistor power amplifiers are definitely with us, attention must be refocused on the pre-amplifier. The main source of distortion in the pre-amplifier is often the tone control circuitry as the power amplifier may require 1V r.m.s. or more to drive it fully, and it usually takes this directly from the output of the tone-control circuit.

The standard one-transistor circuit, as used by A. R. Bailey<sup>1</sup>, gives a total harmonic distortion figure in the region 0.1% to 0.2%. The circuit, adopted by J. L. Linsley Hood<sup>2</sup> is an improvement but necessitates the use of an f.e.t. which is not yet as cheap as a bipolar transistor and, because of its high drain load, requires an output buffer. Ideally a distortion figure in the region of 0.01% at 1V output is desirable.

To achieve this using bipolar devices requires that either the inherent open-loop distortion in the amplifier be reduced, or the open-loop gain increased to give a higher feedback factor for the same closed-loop gain.

The distortion in a transistor with a very high ratio of collector-slope resistance to collector-load is very nearly a function of output current alone. Therefore if the collector load can be raised the output current required to produce a given voltage will be reduced with a consequent reduction in distortion (and an increase in open-loop gain). Unfortunately, the high value of collector load would ordinarily make a high value of supply voltage necessary, and might also make loading effects of the feedback network significant. These difficulties can be overcome simply with an emitter follower performing two functions—providing an output buffer for the high collector load, and giving a bootstrap voltage to raise the effective value of collector load.

The function of bootstrapping is to reduce the actual voltage swing across the collector load resistor for a given collector current required to produce this change, and hence raise the effective resistance of the collector load. This can be achieved by driving the top end of the collector resistor in step with the collector

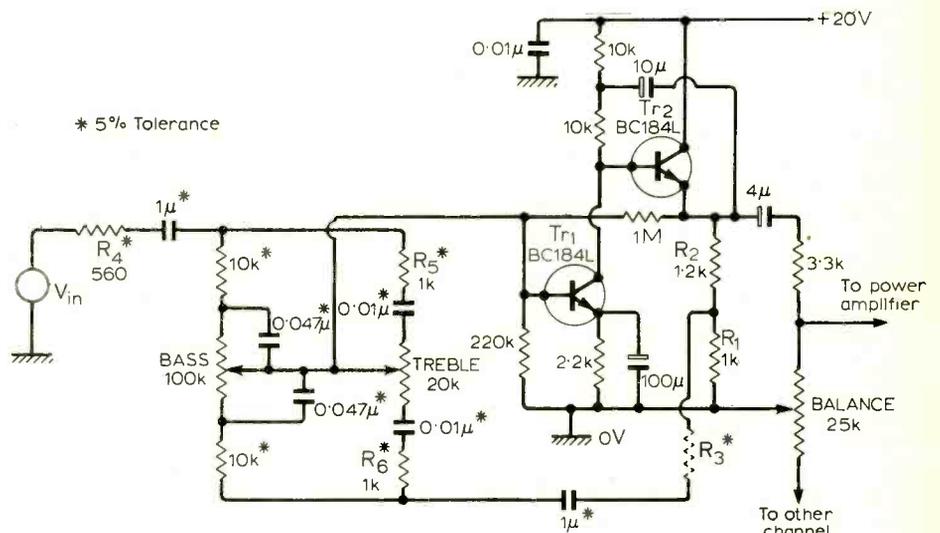


Fig. 1. The complete tone-control circuit built round two n-p-n silicon transistors.

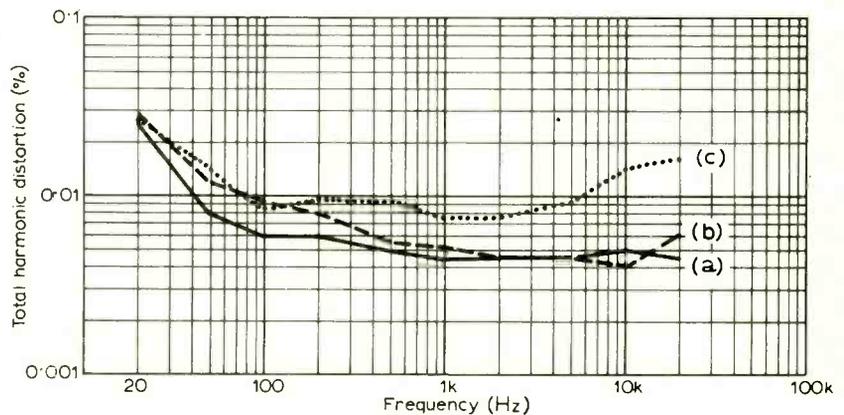


Fig. 2. Total harmonic distortion as measured at 2V output. (a) is the measured t.h.d. of the signal generator, (b) the distortion curve with the tone control flat, and (c) the distortion curve with maximum bass and treble boost.

voltage. The final arrangement is shown in Fig. 1.

The circuit as tested omitted  $R_3$  giving  $R_4$  equal to 560Ω with the values of  $R_1$  and  $R_2$  shown. The gain was 2.2:1 at centre frequency with a subsequent loss of about 1.2:1 with the balance control fitted as shown. The distortion figures for a constant output of 2V r.m.s. are shown plotted against frequency (Fig. 2). The distortion curves for the test oscillator

used and the distortion measured at the output of the amplifier are substantially the same up to 2kHz but, with the treble control set for maximum boost, there is a slight rise at high frequencies. This may have been due to emphasis of the harmonics produced by the test oscillator itself because of the rising characteristic of the amplifier at high frequencies.

The output clips at 6V r.m.s. and with the controls set to the "flat" position, the

\*University of Sussex

total harmonic distortion from 40Hz to 20kHz was measured to be 0.01% or less at 5V r.m.s. output.

The signal-to-noise ratio measured with reference to 1V r.m.s. output over a 20kHz bandwidth was 104dB and the rise time to a step input, 0.1μs.

The circuit may be modified to suit personal taste as required. The relevant equations are as follows:

$$\text{gain} = \frac{R_1 + R_2}{R_2} \quad (1)$$

$$R_1 + R_2 \approx 2k\Omega \quad (2)$$

$$R_3 = R_4 - \frac{R_1 R_2}{R_1 + R_2} \quad (3)$$

Some increase in distortion may result if the gain of the circuit increases beyond 5 or 6 although this may be acceptable especially if the required output voltage is fairly low, as the distortion is a function of output voltage.

Balancing equation (3) is important in order to ensure the controls will be set at their electrical centre when the frequency response is flat. In fact a perfect square wave response cannot be achieved for any setting of the controls unless this equation is balanced.

It should also be noted that the output impedance of the stage driving the circuit is part of  $R_4$  because  $R_4$  is the total source resistance. If this is not taken into account equation (3) will be invalid.

If  $R_4$  is greater than about 500Ω the two 1kΩ resistors  $R_5$ ,  $R_6$  can be omitted. They are included only to limit the ultrasonic gain to prevent instability. It is not advisable to increase  $R_4$  above 2kΩ as the treble control range within the audio range will then be restricted.

The transistor type BC184L was used for this circuit in preference to the more common BC109 because, from experience, the latter type had a tendency to oscillate parasitically due to its collector connected metal can.

In conclusion, this circuit has the advantages of a high output voltage with very low output impedance, negligible distortion and good signal-to-noise ratio.

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1. A. R. Bailey, "High-Performance Transistor Amplifier", *Wireless World*, December 1966.
2. J. L. Linsley Hood, "Modular Pre-amplifier Design", *Wireless World*, July 1969.

# Progress in Air Traffic Control

The first stage of the national air traffic control scheme—code name Mediator—has been introduced at the National Air Traffic Control Service centre at West Drayton, Middlesex. Civil and military radar units operating until recently at Heathrow and serving South-east England have now been closed. With 2,500 movements per day at peak times, increasing at about 10% per year, the new system is needed to increase capacity as well as safety. Work on Mediator was initiated by N.A.T.C.S. in 1962 when it was set up to organize a comprehensive air traffic control system for both civil and military use.

Mediator recognizes radar as the controlling agent whereas in the past radar has been a back-up to 'procedural' control. With it, a whole new range of radar, communications, and automatic data processing equipment is being brought into operation with its associated engineering control, maintenance, power station, and new traffic control techniques. Thinking behind the scheme is similar to that proposed in the early 1960s—see 'Electronics for Mediator' *Wireless World* vol.71, September 1965, pages 426-9—but there have been changes since then, partly as a result of difficulties with equipment.

Difficulties with the computer for flight plan processing have meant postponement of the full implementation of stage 1 but in the words of Michael Noble, Minister for Trade '... This was not a reason for delaying other improvements... not dependent on this particular development.' Improvements include completely new consoles—illustrated on page 181—with bright radar displays and a secondary radar facility, providing controllers with aircraft identification codes superposed on the primary radar display.

The secondary system, of course, works only with those aircraft installed with transponders, at present in the minority. They either have a 64-bit coded transponder—which enables a two-digit route code to be shown on the radar display—or 4096-bit coded transponder which allows aircraft to be identified with two additional decimal digits. Further, some aircraft are fitted with altimeter telemetry equipment, allowing flight level to be shown as well.

Facilities which make up stage 1 of Mediator fall into four main parts—radar outstations, communications links, processing and distribution, and display. The most interesting parts of the system are to do with processing and display, but of course the outstations and communications links are vital and much effort has been devoted to their reliability. Of the long-range primary radar stations at Ash, Ventnor, Lowther and St Annes using 50-cm radars—chosen in preference to the alternative 10-cm radar which would give more precisely defined blips but is susceptible to rain effects—three have dual

aerial heads.

All the secondary radars, co-sited with the primary radars, have dual heads. The main heads have duplicated electronics to give a high degree of reliability and to facilitate maintenance without interruption to the service...

Bright radar displays use a scan conversion technique in which primary video data is written into the storage surface of a conversion tube. This is read with a 1024-line scan many times a second reinforcing the 55-cm display and thus achieving television-screen brightness level. This system differs from other scan conversion systems in the way the secondary radar information is added to the primary. Use of two electron guns—with consequent registration problems—is avoided by sharing gun writing time between the two data. When secondary radar information is available, the normal 'square' scan (equal forward and 'flyback' trace time) for primary information is interrupted and the aircraft designation written on the 256-bit line using digital character generation.

With this system it had been possible to superpose primary and secondary displays to an accuracy of ±1.5mm.

There is also a third kind of information on the display tube—a locally generated map together with range rings and other static information.

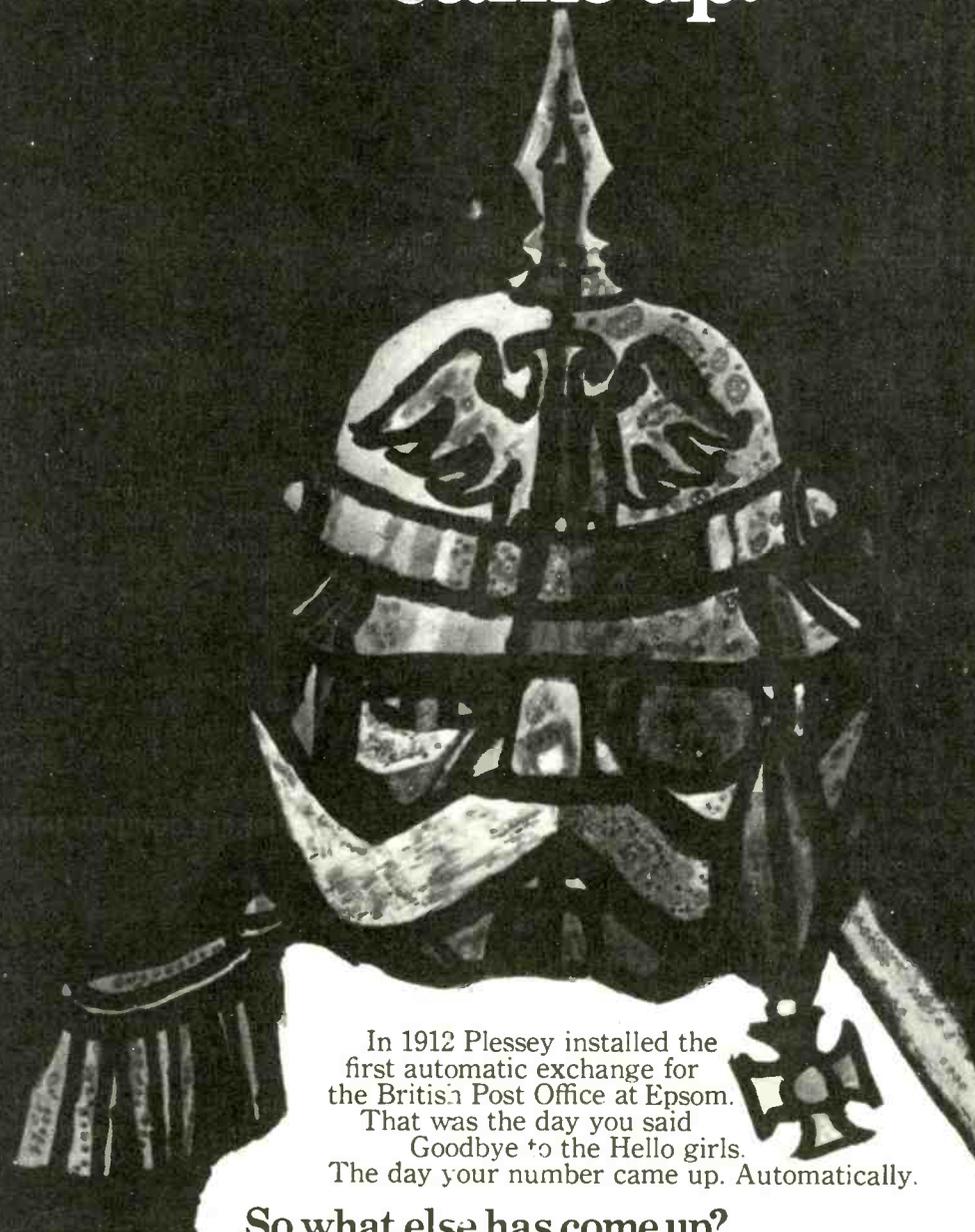
One problem found with this technique of digitally writing the secondary radar information relates to the equal forward and reverse scans. If a display's digit is not accurately matched in position on the forward and reverse scans a jagged sawtooth effect can be produced—an effect which did in fact occur. Attempts to right this by adjusting electrical lengths of cables between scan conversion unit display were unsuccessful and passive delay networks had to be introduced. This needed extra gain from the video amplifiers to maintain display brightness and consequently these are being replaced.

Flight plan processing for controlled airspace is now done with a 32,000-word store using a Ferranti Hermes computer. The system stores flight plans—aircraft identification and certain other information—wind speed and direction, airways structure, link routes, reporting points, and runways in use at Heathrow, verifies the flight plans, calculates an e.t.a. for each reporting point en-route, and prints-out flight progress strips.

For middle airspace a Marconi Myriad computer system will be brought into use by March 1972 at which time R.A.F. middle airspace controllers move to the new centre. This triplicated real-time system was originally planned to be operational by now but software problems led to its postponement. Although only one of the three computers is connected on-line at any time, the others contribute through a kind of self-checking voting

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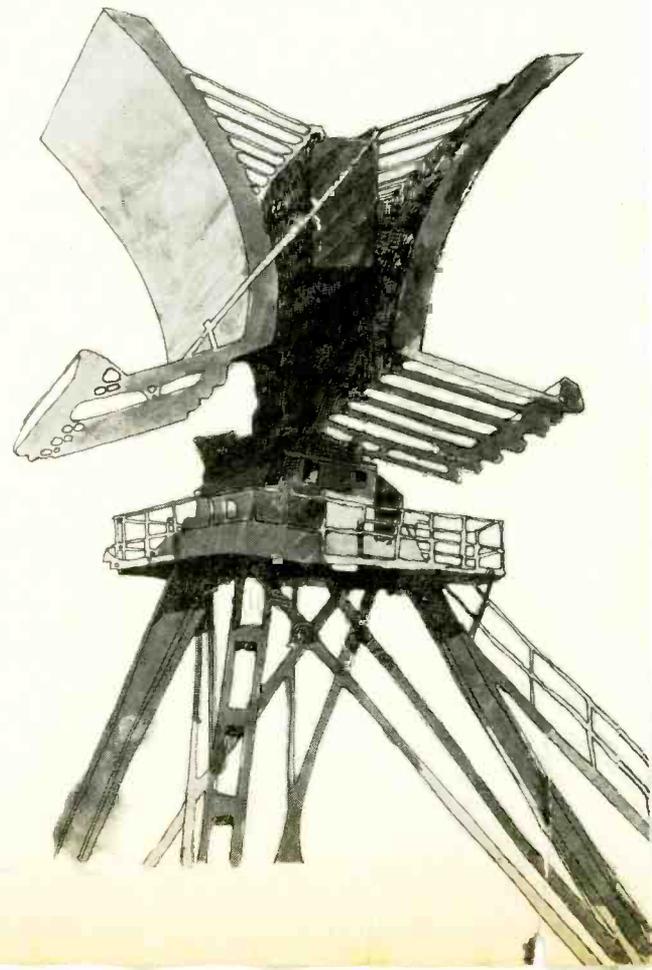
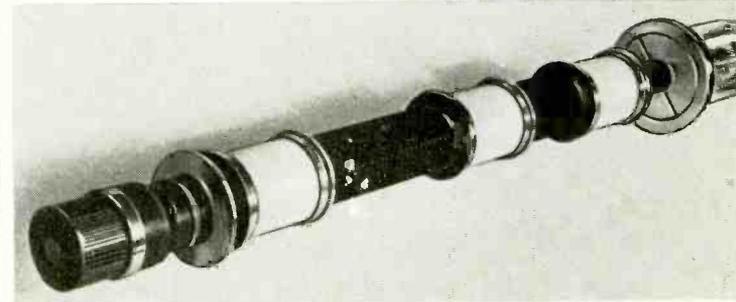
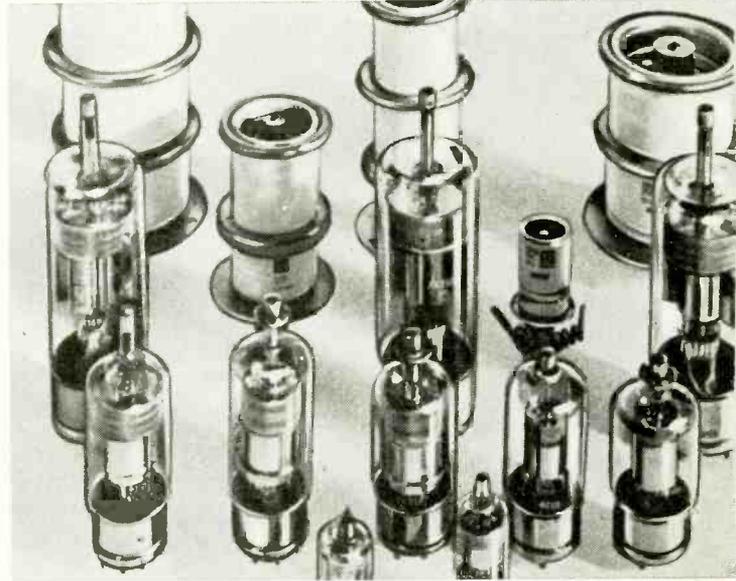
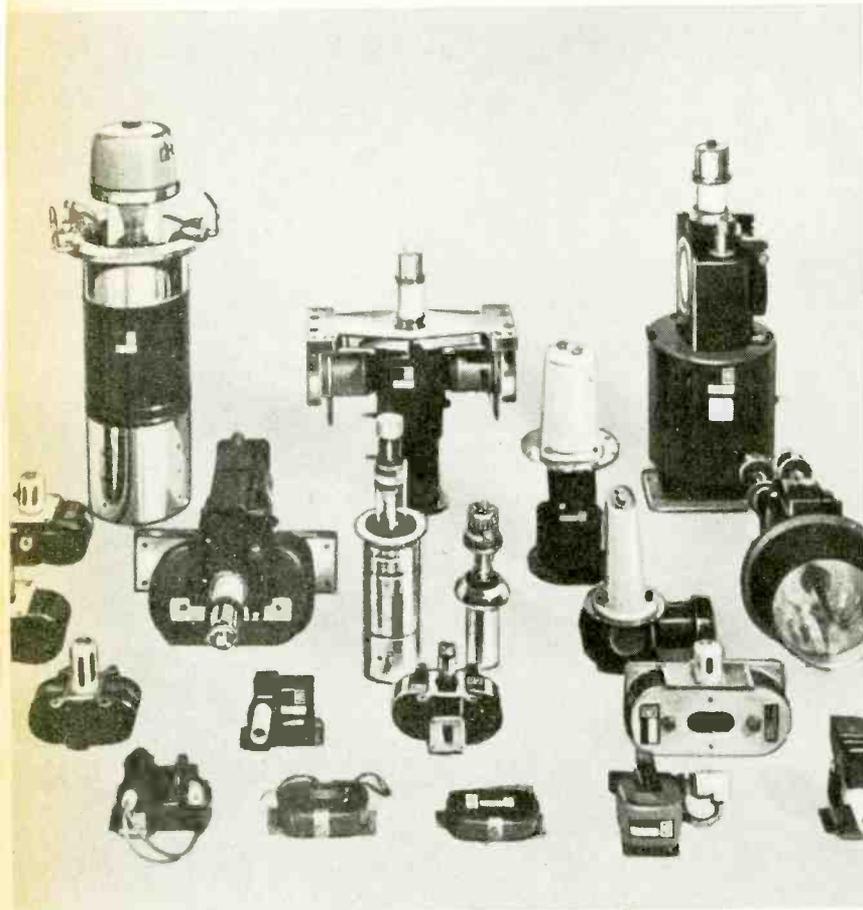
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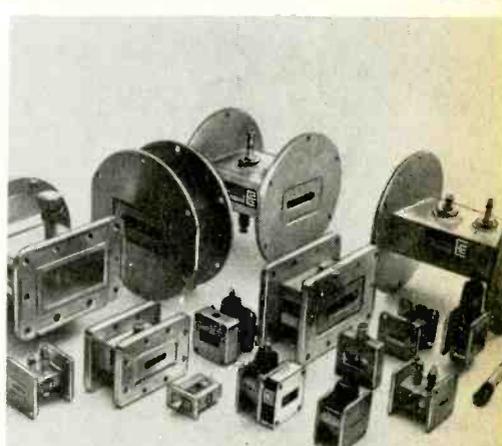
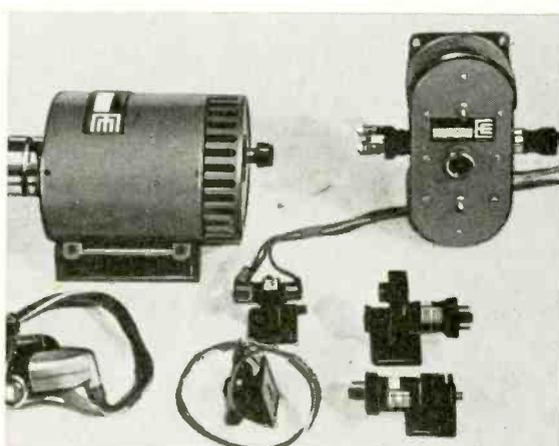
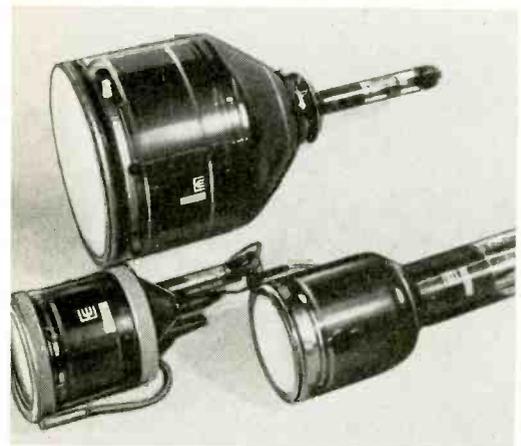
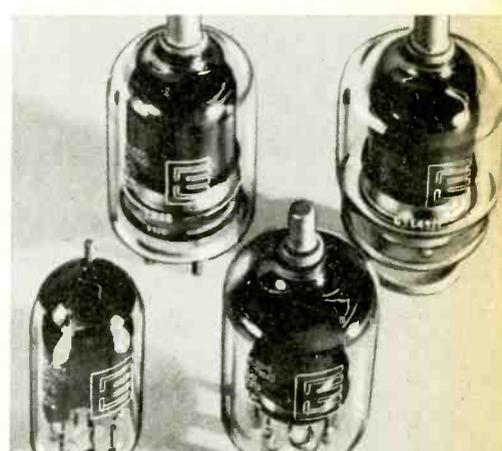
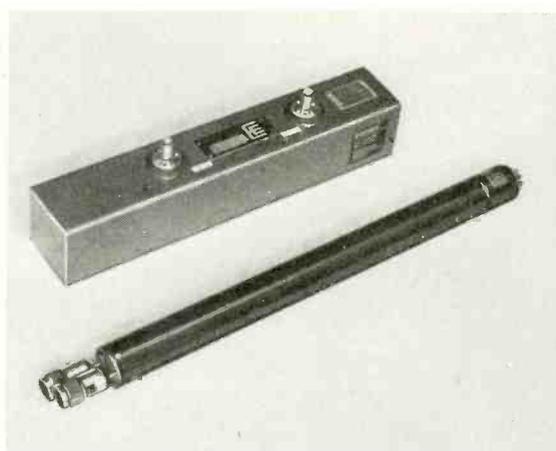
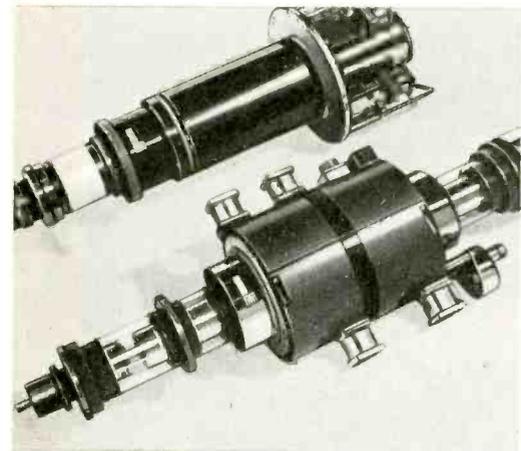
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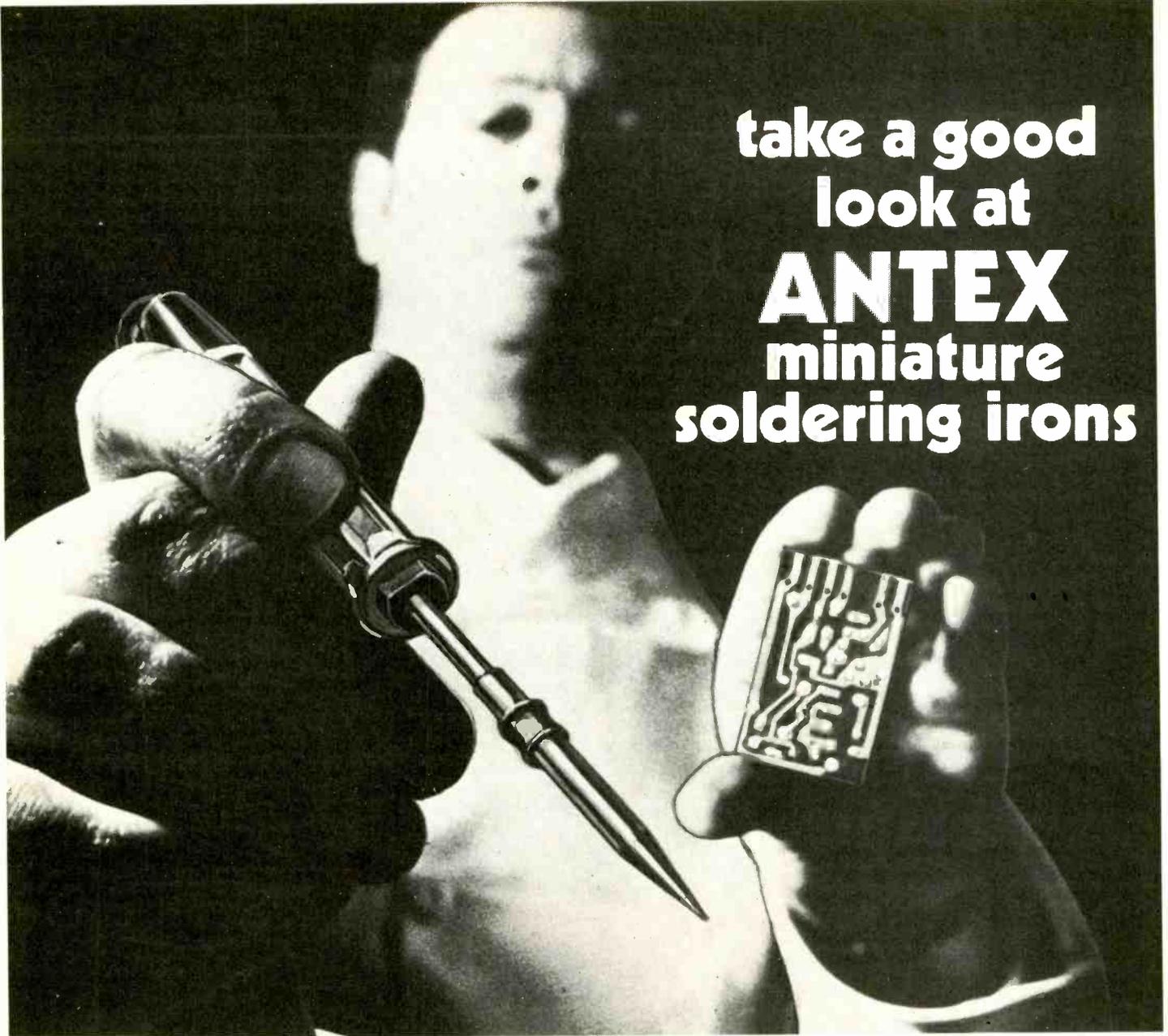
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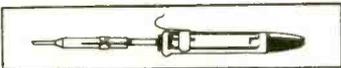
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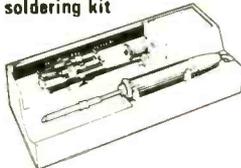


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# Power Amplifier for A.C. Servomotors

A simple general-purpose 10-W design that will drive size-11, -15 or -18 servomotors

by R. J. Wallace\*, M.I.E.R.E. and J. M. Clarke\*, M.Sc.

An amplifier was required for use in the closed loop carrier servo system shown in Fig. 1 to drive the servomotor. The standard operational amplifier would obviously not supply the required amount of power and an investigation of commercially available amplifiers failed to reveal one that could be used in this system.

Amplifiers were found that would drive 20V centre-tapped servomotor windings. However, these were not a lot of use as 20V is not standard for servomotor tacho and reference windings. The reason for this is that suppliers of servo components are closely allied to the aircraft industry, where there is no need for an amplifier to drive the reference and tacho windings, since they are usually driven from the internal 400 Hz aircraft supply. However, in an industrial electronic system, it is not always convenient or desirable to use the 50 Hz line supply for excitation purposes and for the reasons of standardization one amplifier should be capable of driving any winding. The example given is a case in point. Here the inherent property of the a.c. servomotor to remove quadrature signal components is utilized by deriving the reference signal from optical sources. The excitation frequency in such a system can be anywhere in the region of 50 Hz to 1 kHz.

\*Sira Institute

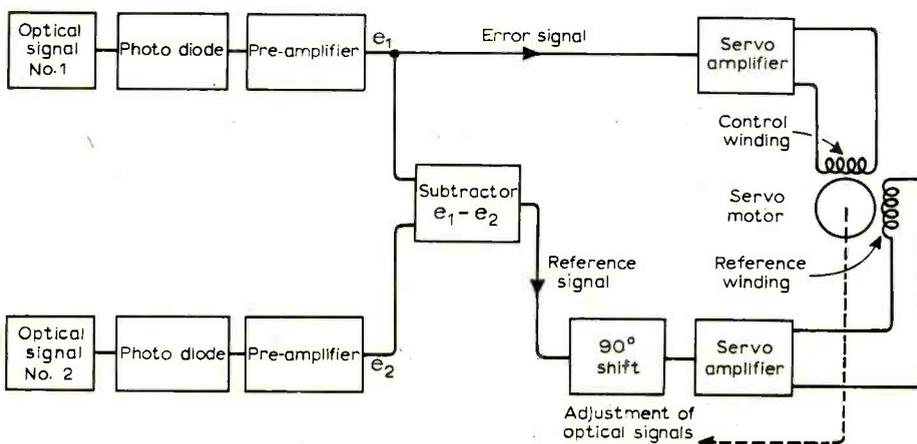


Fig. 1. A closed loop carrier system where an a.c. servomotor is used. Here two optical signals, which are varying at the same time at 440 Hz, are combined to derive a reference signal. In addition one signal is used as the error signal.

An alternative source of power amplifiers and power amplifier designs is the audio field. But here, in order to obtain very low distortion figures, these designs are unjustifiably complicated for industrial application.

In short, there is a need for a simple amplifier, capable of driving any winding of a suitable servomotor which might also be suitable for any application which requires a general purpose a.c. power amplifier.

## Design

It was decided, as discussed, that the amplifier should be able to drive any winding of the servomotor, that all windings should operate at the same voltage and that the amplifier should have no need of the centre tap often available on the control winding.

With regard to power output, one amplifier for reference and tacho windings and one amplifier for the control winding was considered acceptable. The most often used motor is size-11 which dictates that the amplifier should have a power output of 10 W. Incidentally, this is also sufficient to drive single windings on a size-15 or -18 motor.

Since the frequency at which a servomotor may be required to operate can de-

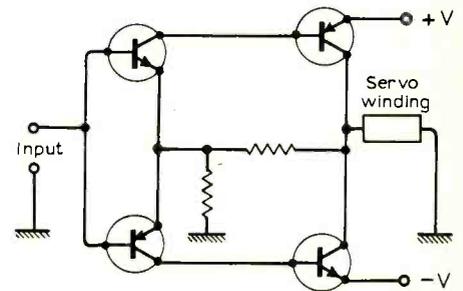


Fig. 2. Basic circuit of the class B output stage employed. Several other possibilities were considered and rejected.

pend on the system as well as the nominal operating frequency of the motor, the frequency response at the amplifier was made as wide as possible, consistent with stability into complex loads, and the final design will drive either 50-, 60-, or 400-Hz motors, and any complex load up to 20 kHz.

The most generally available servomotors have either 26- or 115-V windings. The 26-V version was chosen as this allows a transformerless output stage to be employed. Because the amplifier may be used in systems where the preservation of phase information is important, clipping of the output waveform was arranged to be symmetrical. In addition, no damage will result from reasonable overloads.

Commercial amplifiers often incorporate a 90° phase shift circuit but, since the required phase shift often depends on system phase relationships, such a circuit was not included.

The fact that a.c. servos can tolerate a small amount of distortion suggests the use of a class B, rather than class A, output stage. The higher efficiency of class B means lower heat dissipation and smaller size.

After considering a number of possibilities the simple output stage of Fig. 2 was chosen because of its simplicity, because feedback is easily applied and because the loop gain is low enough to make instability due to complex loads unlikely.

## Circuit details

The circuit of the amplifier is shown in Fig. 3. Each output transistor is driven by a common emitter stage and coupling capaci-

These signals are so close to the 400 Hz commonly used for a.c. servomotors, they can be directly used to excite a servomotor via suitable amplifiers. However to find a source of supply of such amplifiers is a problem.

tors are avoided by the use of complementary circuitry. Overall feedback is by  $R_7, R_8$  and  $R_9, R_{10}$ , with a separate path for each 'half' of the amplifier to assist in stabilizing the bias currents.

The high frequency response is restricted to 23 kHz by  $C_1$  and  $C_2$ . The bandwidth can be increased to about 400 kHz by removing  $C_1$  and  $C_2$  if due care is taken with the component layout. Diodes  $D_1$  to  $D_4$  are for bias current temperature compensation, the final value of the output stage bias being set by  $RV_1$  and  $RV_2$ . All the transistors and diodes are mounted on a common heat-sink which should have a thermal resistance of less than  $2.5^\circ\text{C}/\text{W}$ .

Fuses  $F_1$  and  $F_2$  provide protection against damage when the amplifier is used into a load below the permitted minimum, or short circuits of small duration. They do not give full protection and the provision of additional safeguards was not considered to be worth while. The clipping action protects the amplifier when it is overdriven.

**Performance**

Three amplifiers were built and subjected to the tests detailed below. The amplifier load resistance was  $68\Omega$  and the signal frequency was 400 Hz.

**Voltage gain versus ambient temperature:** The gain of the three amplifiers was measured for 10 V r.m.s. output. The results were as follows.

Amplifier	Gain	
	20°C	70°C
1	5.72	5.85
2	5.72	5.9
3	5.72	5.84

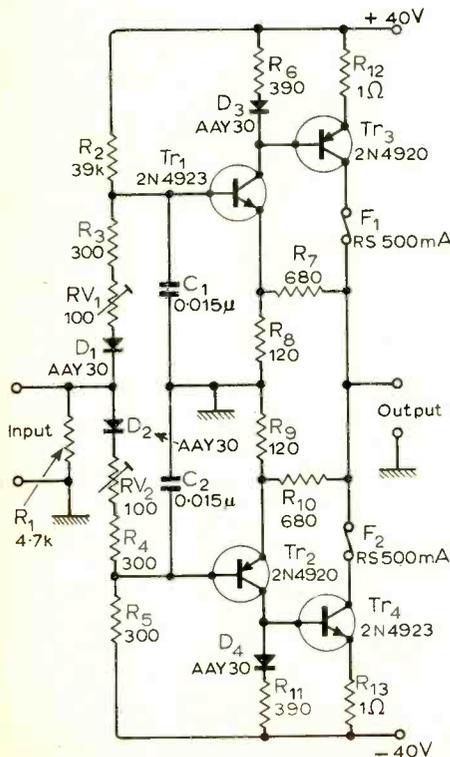


Fig. 3. Final circuit. With the input shorted to ground  $RV_1$  and  $RV_2$  should be adjusted for 1mA quiescent current through  $Tr_3$  and  $Tr_1$ .

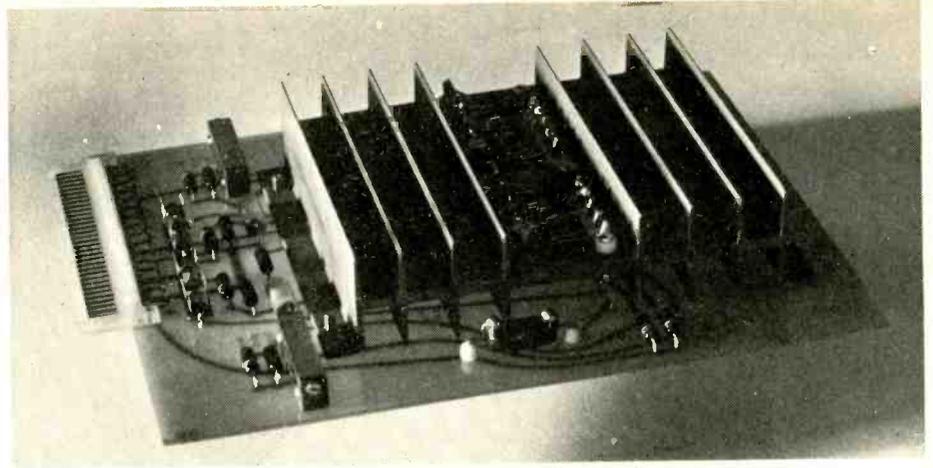


Fig. 4. The completed prototype.

**Clipping levels:** The r.m.s. output voltages at the clipping points of the three amplifiers were measured and are set out below. These voltages are marginally higher than might be expected because of the small amount of distortion.

Amplifier	Clipping voltage (r.m.s.)	
	20°C	70°C
1	28.9	29.2
2	28.7	29.3
3	28.8	29.2

**Frequency response:** The high frequency response has been limited as described earlier and is 3 dB down at 23 kHz. As the amplifier is d.c. coupled, the minimum frequency for full output is determined by the maximum period over which it is permitted to average the instantaneous dissipation. While 40 Hz has been specified as the minimum, a margin of safety has been allowed for constructional and device parameter variations. One amplifier, has in fact operated satisfactorily down to 30 Hz.

**Input impedance:** The input impedance is slightly dependent on level, the lowest value being 3.5 kΩ.

**Quiescent current stability:** Since the diode and transistor  $V_{be}$  characteristics are not matched perfectly, some change in bias current is observed with change in temperature. The figures were as follows:

Amplifier	Bias current in mA			
	20°C		70°C	
	$Tr_3$	$Tr_4$	$Tr_3$	$Tr_4$
1	1	1	26	25
2	1	1	28	28
3	1	1	27	26.2

The change of bias current with temperature is determined by the type of compensating diodes and the current through them. Since it is not practically possible to achieve a constant bias current with respect to temperature, a positive temperature coefficient is accepted so that increase of temperature will not lead to distortion because of insufficient bias current. Typically the bias current falls to 0.1mA at 0°C but the major change, to approximately 27mA at 70°C, occurs between

50°C and 70°C. The figure of 27mA represents only about 5% of the peak current at full output.

**D.C. current in the load:** The amplifier is completely d.c. coupled and any d.c. appearing at the input, perhaps, due to offset errors in a preceding operational amplifier, for example, causes a d.c. current to flow in the servomotor. In the worst case approximately 60 mA at the input causes 5 mA to flow in a  $68\Omega$  resistive load.

Since completion of this amplifier (Fig. 4) integrated circuit audio power amplifiers have become available in the U.K. with power outputs up to 5 W. Perhaps one of the integrated circuit manufacturers would think it worth while to market a servomotor-amplifier with a similar performance to that described by the authors.

# Announcements

**Duty free.** All approved v.h.f. multi-channel radio communication and navigation equipment used in light aircraft is now exempt from U.K. import duty.

The entire share capital of **General Video Systems Ltd.**, the main U.K. distributor of Shibaden broadcast and c.c.t.v. equipment, has been purchased by Shibaden. The name of the company has been changed to **Shibaden (U.K.) Ltd.**, which will continue to trade from 61/63 Watford Way, Hendon, London NW4 3AX. Tel: 01-202 8056.

Nortronics Company Inc, of Golden Valley, Minnesota, manufacturers of magnetic recording heads, have formed **Nortronics, S.A.** in Brussels to market their products in Europe. A manufacturing facility will be formed at a later date.

The Avionics Division of Plessey Electronics Group are to supply **IFF shipborne transponders and auto-decoders** valued in excess of £½M to the Royal Navy.

The Canadian Department of Transport has awarded a £118,000 contract to Decca Radar (Canada Ltd, for a **harbour radar system** to cover the Chedabucto Bay area, Nova Scotia.

The Scottish Northern Lighthouse Board has purchased four **radio beacons and ancillary equipment**, valued at £25,960, from AGA (U.K.) Ltd.

# High-gain Audio Voltage Amplifier

by D. Leblebici\*

One of the commonly used feedback amplifier circuits is the 'feedback pair', where feedback is applied from the collector of the second transistor to the emitter of the first transistor (Fig. 1(a), ref. 1).

The feedback circuit described here is a modified form of the conventional feedback pair (Fig. 1(b)). Feedback is the series voltage type applied from the emitter of  $Tr_3$  to the emitter of  $Tr_1$ . The circuit has some advantages as compared to the conventional feedback pair:

- The output as well as the input terminals of the circuit are outside the feedback loop and consequently the amount of feedback is independent of the source and load impedances.
- The input and output signals are in phase opposition and as a consequence it is possible to apply a second feedback loop (parallel voltage feedback) from the collector of  $Tr_3$  to the base of  $Tr_1$ .

\*Elektrik Fakültesi, Teknik Üniversite, İstanbul.

<sup>1</sup>National Bureau of Standards, preferred circuit no. 201.

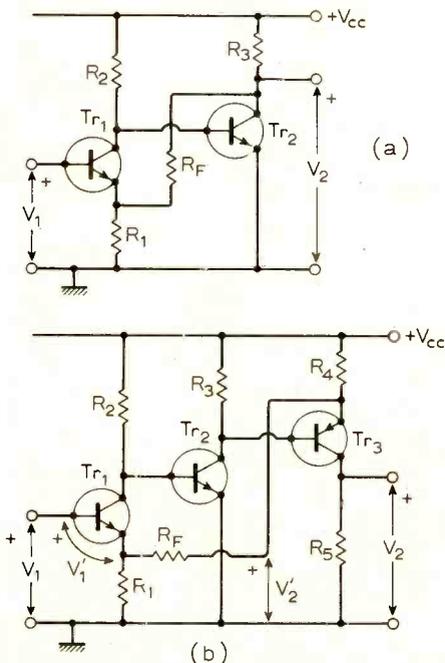


Fig. 1. Comparison of feedback pair (a), and modified form of feedback pair (b). One advantage of the modified form is that feedback is independent of impedance.

● The feedback voltage is taken from the emitter of  $Tr_3$ . The negative feedback acts to decrease the distortion of the voltage wave at that point. The relation between the output voltage  $V_2$  of the amplifier and the voltage  $V_2'$  fed back can be written as

$$\frac{V_2}{V_2'} \approx \frac{R_5 I_{e3}}{R_4 I_{e3}}$$

provided that  $R \gg R_4$ . For a high-gain transistor  $I_{e3}/I_{e3}$  is very close to unity. Hence

$$\frac{V_2}{V_2'} \approx \frac{R_5}{R_4}$$

This shows there is an additional and practically linear (low distortion) voltage gain of magnitude  $R_5/R_4$  from the emitter to the collector  $Tr_3$ .

As the voltage gain from the collector of  $Tr_2$  to the emitter of  $Tr_3$  is approximately equal to unity, the voltage gain from the input terminal to the emitter of  $Tr_3$  must be equal to the gain of a conventional feedback pair using the same transistors  $Tr_1$  and  $Tr_2$  and the same circuit components  $R_1, R_2, R_3$  and  $R_F$ . As this gain is approximately equal to  $(R_F + R_1)/R_1$  (ref. 2), the total voltage gain becomes

$$A_v = \frac{V_2}{V_1} \approx \frac{R_F + R_1}{R_1} \cdot \frac{R_5}{R_4}$$

The only drawback of the circuit is its relatively high output resistance, which is approximately equal to  $R_5$ .

### Experimental circuit

The experimental circuit diagram is shown in Fig. 2. The stages are directly coupled. To stabilize the quiescent points a d.c. feedback across the first two transistors is used. A collector current of about  $200 \mu A$  is chosen for  $Tr_1$  this being the optimum collector current of transistor BC109 for minimum noise. The transistor operating points and component values have been calculated for a sufficiently high open-loop gain and as high a dynamic range as possible.

The calculated open loop gain  $V_2'/V_1$  is 11600, a value that is sufficiently high. The additional gain provided by  $Tr_3$  is about 10.

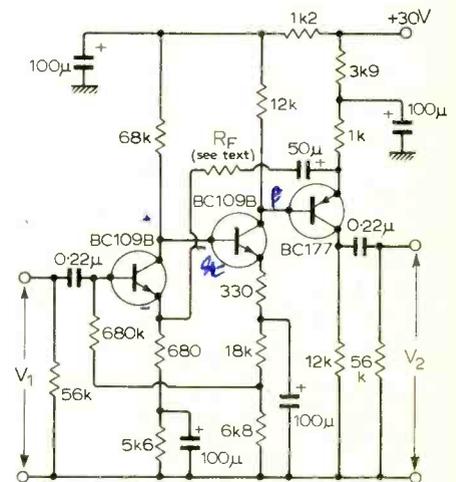


Fig. 2. Practical circuit of preamplifier with voltage gain of 100 for  $R_F = 5.6 \text{ k}\Omega$  or 1000 for  $R_F = 78 \text{ k}\Omega$ . Total harmonic distortion is 0.02% and 0.04% respectively.

The measured overall voltage gain for  $R_F = 5.6 \text{ k}\Omega$  was  $A_v = 100$  (calculated value:  $A_v = 92$ ) and the available maximum output swing was 18 V pk-pk (6.4 V r.m.s.). For an output voltage of 5 V r.m.s. the total measured harmonic distortion was 0.02%. For  $R_F = 78 \text{ k}\Omega$  the measured  $A_v$  was 1000 (calculated value:  $A_v = 1150$ ). The maximum output voltage was again 18 V pk-pk and the total harmonic distortion was 0.04% ( $V_2 = 5 \text{ V}$  r.m.s.). The measured lower and upper cut-off frequencies for both cases were 17 Hz and 200 kHz.

Consequently, the circuit is very convenient as a high-gain audio preamplifier. The possibility of applying a second, independent, parallel voltage feedback loop makes it possible to use the circuit as a low output impedance, moderate gain and high dynamic range booster amplifier. With frequency dependent feedback, it is also possible to use the circuit as a low distortion equalizer amplifier.

<sup>2</sup>Millmar & Holkias, 'Electronic devices and circuits', pp. 502, 3.



# Electronic Building Bricks

## 11. Information paths between units

by James Franklin

We have seen that information—which might be numbers of objects in an electronic counting system or light intensity in a television system—may be represented by electrical variables. We have also looked fairly closely at some of these variables, and seen how they exist in circuits (Part 5). We have not, however, studied how the electrically represented information is actually conveyed from one electronic unit to another.

Part 1 showed block diagrams of a television set and a computer and explained that the lines joining the blocks indicate paths for information. Some readers may have thought it odd that these paths, as well as the functional units themselves, were considered as “building bricks”. The justification for this idea is that in practice these information paths are provided by *circuits*—particular arrangements of conductors.

At this point we should get to know one of the conventions of electronic diagrams—conventions, incidentally, which are automatically understood all over the world. The electrically represented information which comes out of an electronic unit is called the *output*, and is normally shown emerging from the right hand side of the unit. Conversely, the information an electronic unit receives is called the *input*

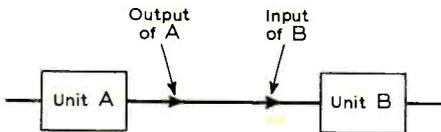


Fig. 1. The convention of a single line used to indicate an information path, the output of one unit becoming the input of another. (This line is not an electrical conductor as such.)

and is shown going into the left hand side of the unit. Thus, in Fig. 1 the output of unit A becomes the input of unit B.

As we have said the basic means by which the information is conveyed is electrical energy—more specifically the rate of delivery of energy, which is power (Part 8). In practice, however, we don't usually consider power as representing the information, mainly because it is not very convenient to detect and measure the varying power that flows from unit A to,

unit B. The measuring instruments commonly used in electronics respond to other electrical variables, variables that are proportional to the power, in particular potential difference and current. This complicates the picture, because in any circuit where electrons are moving there must be an e.m.f., and this creates both potential difference and current. Which one, then, represents the information?

The answer is, simply, whichever one the electronic designer has chosen to represent the information. The other variable is there as well, in the sense that it can be measured, but it is not taken as significant as a bearer of information.

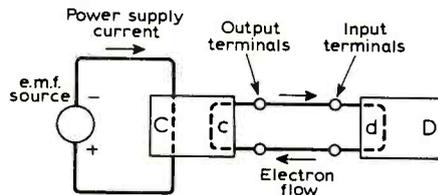


Fig. 2. An electric circuit joining two units so that it provides an information path between them. In this circuit, current is the significant variable.

As an example, Fig. 2 shows the basic principle by which information may be transmitted from one unit to another using current as the significant variable. It can be seen that there is a complete electrical circuit passing through part of unit C and part of unit D. Electrons are made to flow in this circuit by an e.m.f. existing in section c of unit C. (We will not discuss exactly how this e.m.f. comes to be there, as we are not at the moment concerned with the actual functions of units C and D, but it originates from the e.m.f. source supplying electrical power to unit C.) The electron flow rate (current) varies with time and this variation represents information (Part 2). In unit D, at section d of the circuit, there is a means of continuously detecting or responding to the value of the current. Thus the information represented by the current variation is conveyed into D.

It would be possible to detect a potential difference (resulting from the e.m.f. at c) between the two output terminals of unit C, or between the two input terminals of unit D, but this in itself is not significant

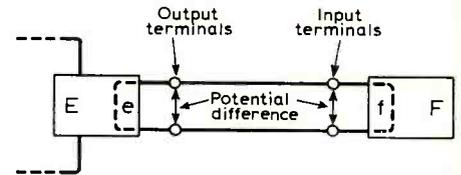


Fig. 3. Another circuit providing an information path between units, but using potential difference as the significant variable.

because current has been chosen as the information bearing variable.

Fig. 3 shows how information may be transmitted using potential difference (p.d.) as the significant variable. Here again there is a circuit completed through the two units. There is a p.d. between the output terminals of unit E, resulting from the e.m.f. existing at e, and it is the variation of this p.d. which represents the information. The input terminals of F, being connected to the output terminals of E, have the same p.d. between them (analogy: the water level in a water gauge is the same as the water level in the tank to which it is attached). This p.d. is present at f, where there is a means of continuously detecting it, so that the information represented by the variation of p.d. is conveyed into F.

Again there is a current flowing in the circuit, but here it is this current which is not significant as an information bearer. In some electronic systems designed to use p.d. as the significant variable the current flowing is extremely small, for example less than a millionth of an ampere. In such cases one can think of the information virtually as being conveyed by a variation of p.d. alone. In other cases the circuit is not completed at f (imagine broken line absent), and unit F detects the information as an electric field (Part 5) which is created by the p.d. between the input terminals.

As a practical example, E and F in Fig. 3 could be two stages of an electronic amplifier (Part 9). The potential difference at e would be that developed across a load in the high-power circuit of stage E, while circuit f would be the low-power control circuit of stage F.

# Personalities

**D. B. Weigall, C.B.E., M.A., F.I.E.E.**, retires from the position of deputy director of engineering at the end of March after more than 37 years of service with the B.B.C. A graduate of Christ Church, Oxford, he joined the Corporation in 1933. After three years in the Research Department he became assistant to the superintendent engineer, studios. From 1940 to 1942 he was seconded as chief engineer to the Malaya Broadcasting Corporation and from 1943 to 1946 he was technical adviser on broadcasting to the Ministry of Information. After his return to the B.B.C. Mr. Weigall joined the Planning and Installation Department in 1948. He was appointed chief engineer, External Broadcasting, in 1962; assistant director of engineering in 1963 and deputy director of engineering in 1967.

**G. Stannard, B.Sc., F.I.E.E., A.C.G.I.**, retires on May 24th from the position of chief engineer, communications, in the B.B.C. Educated at the City & Guilds College, London University, he joined the B.B.C. in 1932 and after working in the London control room and the recording section he transferred to the Lines Department in 1935. He has been chief engineer, communications, since 1965. In this position he has been responsible for the planning and commissioning of the vision, sound and communication networks used for the distribution of programmes and the provision of other communication facilities.

As a result of the retirements of D. B. Weigall and of G. Stannard, the B.B.C. has announced the following appointments:

**D. E. Todd, B.Sc.(Eng.), F.I.E.E.**, at present assistant director of engineering, will assume the responsibilities of deputy director, in addition to his present responsibilities for the engineering specialist departments. Mr. Todd joined the B.B.C. in 1946 and was head of Transmitter Planning and Installation Department from 1965 until his appointment as assistant director of engineering in 1968.

**T. B. McCrerrick, F.I.E.E.**,

**F.I.E.R.E.**, will become assistant director of engineering and will be responsible for the Transmitter and Communications Departments, the Engineering Information, Engineering Training and Engineering Personnel Departments. Mr. McCrerrick joined the Corporation in 1943 and transferred in 1949 to the television service, where he later held the posts of engineer-in-charge, television studios, and head of engineering, television recording. Since last June he has been chief engineer, radio broadcasting.

**D. R. Morse, F.I.E.E.**, lately chief engineer, capital projects, has been appointed to the new post of chief engineer, networks and communications. He will have special responsibilities for negotiations with the Post Office and the Ministry of Posts and Telecommunications regarding the provision of B.B.C. programme and communication networks.

**J. D. MacEwan, B.Sc., F.I.E.E., M.I.E.R.E., A.Inst.P.**, becomes chief engineer, radio broadcasting. Mr. MacEwan joined the B.B.C. in 1947. He was appointed a senior lecturer at the Engineering Training Centre, Evesham, in 1956. Since August 1969 he has been chief engineer, regions.

**Tony Martin**, who joined International Rectifier in 1959 when they commenced manufacture at their Oxted, Surrey, plant, has been appointed Northern European sales manager. He worked for several years as an application engineer and then transferred to the marketing department where he held successive posts of sales office manager and product sales manager. He will be responsible for the industrial sales organization in all Northern European countries.

**James Lionel West**, aged 35, has been appointed manufacturing manager of the m.o.s. division of Emihus Microcomponents Ltd at Glenrothes, Fife. Mr. West's previous appointments were with General Instruments, Elliott Auto-

mation and the A.E.I. research laboratories at Harlow.

**P. Scargill, M.I.E.R.E.**, general manager of the Electronics Division of Union Carbide U.K. Ltd, is transferring to Union Carbide Europe, in Geneva, where he will be responsible for the group's European electronics business. Mr. Scargill, who is 36, joined Union Carbide in 1966 from Hughes International (U.K.) and was previously U.K. capacitor sales specialist with International Electric Co. of New York.

**Tony Wynter**, who recently joined Devices Instruments Ltd, of Welwyn Garden City, Herts, has been appointed managing director of Devices Pty. Ltd, the sales and service organization set up in Sydney, Australia. Before joining Devices Mr. Wynter was a member of the medical electronics staff at the National Hospital for Nervous Diseases, Queens Square, London.

**Peter D. Simmons** has joined SE Laboratories as product manager of their new digital instrumentation division. Immediately prior to joining SE Labs he was sales manager for Racal, and before that was with Solartron and Dowty.

**Guy Barnes, Ph.D., B.Sc.**, has joined Emihus Microcomponents Ltd as technical manager. Dr. Barnes, aged 38, graduated in physics and mathematics at Reading University. He then undertook on behalf of the Admiralty four years research at



Dr. Guy Barnes

the University into the surface properties of semiconductors. He joins Emihus from Texas Instruments where he was integrated circuit department manager. From 1958 to 1963 he was chief physicist with Mining and Chemical Products Ltd.

**A. G. Touch, M.A., D.Phil.**, who is 60, is retiring from the post of chief scientist at the Government Communications Headquarters. A graduate of Jesus College, Oxford, Dr. Touch joined Watson-Watt's radar team at Bawdsey research station in 1936. For his

contributions to the development of meter-wave AI and ASV he received a substantial award on the recommendation of the Royal Commission on Awards to Inventors. From 1941 to 1947 he was liaison officer with the British Joint Services Mission in Washington. On his return to this country he became superintendent of the Blind Landing Experimental Unit at Martlesham Heath, Suffolk, and from 1952 to 1954 was deputy director of electronics research and development (air) in the Ministry of Supply. From 1954 to 1959 he was director of electronics research and development (ground), M.o.S., and then senior superintendent of the Radio Department at R.A.E., Farnborough. He is to be succeeded as chief scientist at the Government Com. H.Q. by **Ralph Benjamin, Ph.D.**, who is at present director of Admiralty Underwater Weapons Establishment at Portland. Dr. Benjamin, who is 47, joined the Admiralty Signal Establishment in 1944. His particular fields of research at A.S.E. were pulse techniques and weapon control. Dr. Benjamin's place at Portland is being filled by **G. L. Hutchinson, Ph.D.**, at present head of the Military and Civil Systems Department at the Royal Radar Establishment, Malvern. A graduate of King's College, London, Dr. Hutchinson joined the Scientific Civil Service in 1939 working on the installation of the coastal radar chain. In 1943 he joined the Telecommunications Research Establishment, then from 1948 to 1954 was at R.A.E. Farnborough and later on the staff of the British Joint Staff Mission in Washington.

Siliconix Ltd., of Swansea, recently announced two new appointments. **David Thomson**, has joined the company as sales engineer. In 1963 he joined A.E.I. Telecommunications as a student apprentice concentrating on the design of linear and digital telecommunication equipment and gained an honours degree in electronic engineering at The City University, London. In 1968 he went to Elliotts, Rochester, as a development engineer on airborne digital computers. **David J. West**, aged 23, who graduated in electrical engineering only this year from the University College of Swansea, has joined the company as applications engineer.

## OBITUARY

**Leonard Walter Fillmore**, managing director of Jackson Brothers (London) Ltd, which with his father and brother he founded in the early 'twenties, died on January 30th aged 67. During the war years he was very active on Ministry Standardization Panels and right up to the time of his death he worked on the Variable Capacitor Standardization Panel of R.E.C.M.F.

# New Products

## Beam tetrodes

Two compact conduction-cooled beam tetrodes are available from M-O Valve Co. Typically they give 400W output in f.m. service up to 175MHz, and 200W up to 500MHz. In s.s.b. service 300W p.e.p. output is obtainable up to 175MHz. Both tubes are electrically identical but differ in construction. The CCS1 has a square copper block fitted to the anode which is intended to be bolted to a heat sink directly or by means of a beryllia heat conducting block when electrical isolation is required. The CCS2 incorporates an electrically isolated flange intended to be clamped directly to a heat sink. Thermal resistance between anode and flange is made low by the use of beryllia ceramic. There is no significant increase in output capacitance. These conduction cooled tubes are fully replaceable in new equipment for existing well established forced air cooled types. The conduction cooling assists circuit designers by eliminating moving parts in the cooling system and so achieves greater reliability and compactness at reduced power consumption. M-O Valve Co. Ltd, Brook Green Works, London W.6. WW315 for further details

## R.M.S. voltmeter

Model A130 r.m.s. voltmeter from Prosser Scientific Instruments allows waveforms from d.c. to 100kHz with crest factors (peak to r.m.s. ratio) up to 10:1 to be analysed. An averaging time constant facility allows integration from 1 to 300s. At very low frequencies it is sometimes useful to measure the instantaneous power by measuring the instantaneous square of the voltage. This facility is provided on the A130 with a time constant of 10ms. The instrument, designed for bench use or rack mounting, employs a large meter display ( $\pm 2\%$  accuracy) as



well as providing an output for higher accuracy readings. The specification includes the following:

input impedance	1M $\Omega$ 25pF
overload	300V d.c. or peak on ranges 1mV-10V. 1000V d.c. or peak on ranges 30V to 300V.
output	4 inch panel meter scaled 0-10V, 0-3V, and -15dB to +2dB. Linear d.c. output 0-10V and 0-3V full scale depending on range selected.
output impedance	< 10 $\Omega$ . 10mA max.
accuracy	$\pm 0.5\%$ of full scale.
power requirements	115/240V a.c., 50/60Hz, 50VA.
dimensions	445mm $\times$ 133mm $\times$ 344mm.
weight	5kg.

Prosser Scientific Instruments Ltd, Lady Lane Industrial Estate, Hadleigh, Ipswich, Suffolk.  
WW317 for further details

## Miniature silicon bridge rectifiers

A range of 1.2A silicon bridge rectifiers is available from General Instrument (UK). The assemblies are of flat construction with in-line leads, and measure 23.5  $\times$  4mm. Called the FB series they have a p.i.v. rating of up to 600V and a one cycle surge capability of 50A at operating temperatures from  $-55^{\circ}\text{C}$  to  $150^{\circ}\text{C}$ . General Instrument (UK) Ltd, Stonefield Way, Victoria Road, South Ruislip, Middx.

WW314 for further details

## Numeral indicator tubes

ITT Components Group Europe, by arrangement with Burroughs Corporation, are manufacturing two new long-life side-view cold-cathode indicators. The 5853S is designed for use in time-sharing by anode strobing applications; the 5870S is intended for d.c. and pulsed operation with peak cathode currents up to 10mA. Both tubes have a display of numerals 0 to 9 inclusive with one decimal point on the left and right of each main character, and a character

display area of 13.5mm (height)  $\times$  7.6mm. Maximum overall height is 30.5mm, and diameter 13mm. Base connections are 14 tinned leads with in-line configuration for printed circuit or socket use. Typical operating conditions are:

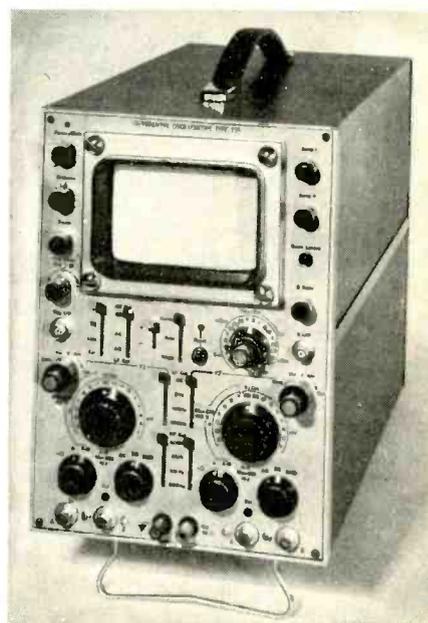
type 5853S	
anode supply voltage	200V
peak cathode current	14mA
pulse duration	100 $\mu\text{s}$
pulse repetition frequency	500Hz
type 5870S	
anode supply voltage	200V
cathode current (with decimal point)	3.2mA
cathode pre-bias voltage	67V

ITT Components Group Europe, Valve Product Division, Brixham Road, Paignton, Devon.

WW313 for further details

## Twin-channel oscilloscope

Differential oscilloscope type 155 from Bradley Electronics has an input impedance of 100M $\Omega$  shunted by 1pF, and 100dB common-mode rejection. The common-mode signal may be as high as  $\pm 15\text{V}$  at maximum sensitivity. Sensitivity is 100 $\mu\text{V}/\text{cm}$ , but this can be increased to

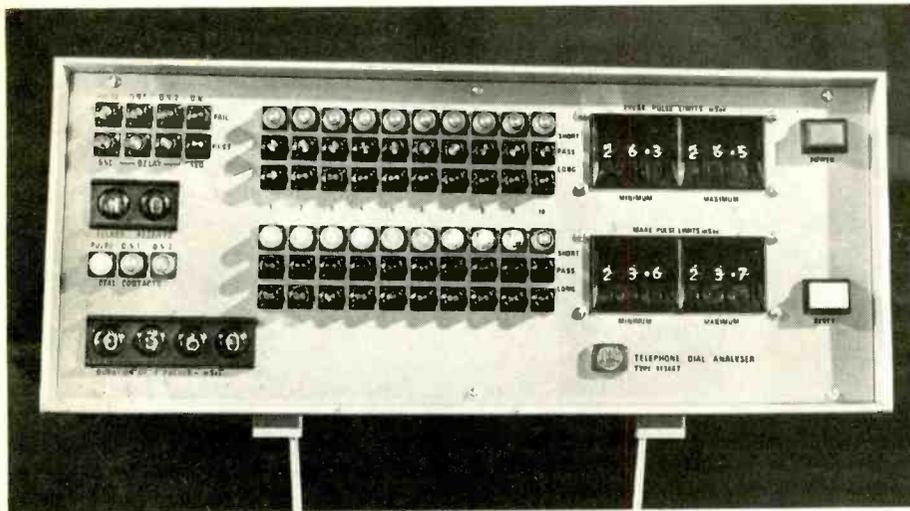


10 $\mu\text{V}/\text{cm}$  by cascading the two channels. The timebase will operate down to 5s/cm. Outputs suitable for driving pen recorders are provided and the oscilloscope can be used as an X-Y plotter. Trigger arrangements include auto and single shot and the general facilities provided include beam locate and internal calibration. G. & E. Bradley Ltd, Electral House, Neasden Lane, London N.W.10.

WW 303 for further details

## Telephone dial testers

An instrument which tests telephone dial pulsing—in terms of operate and release times of individual dial pulses—is made by Amalgamated Wireless (Australasia) Ltd. Traditionally, dial performance during



production adjustment is done by measuring pulsing speed and make-to-break ratio on an average basis. But the telephone dial pulse monitor type IT1466 compares make and break duration of each pulse with preset tolerance limits. Indication is given if any interval is outside the limits set by the operator by lighting a 'short', or 'long' lamp. In addition the instrument detects excessive contact bounce, counts the number of pulses received, and checks sequence and relative timing of off-normal contact re-closure. Pulse analyser type IT1467 performs similar tests and additionally gives 'short', 'pass' and 'long' indication for each of ten contacts. Both incorporate a crystal oscillator timing standard. Available in the U.K. from Amalgamated Wireless (Australasia) Ltd, Aldwych House, 81 Aldwych, London W.C.2.  
**WW311 for further details**

### Capacitors for s.c.r. commutation

A range of capacitors for high-power controlled rectifier circuits is made by Aerovox Corp. Developed to provide low inductance and low series resistance needed to turn off s.c.r.s, the capacitors are said to be made more reliable than general-purpose types. They are made with three kinds of dielectric—paper, metallized paper and polycarbonate—and are designed to dissipate internally generated heat. Paper dielectric gives lowest cost, but as might be expected gives largest bulk. Size is reduced by using metallized paper, but cost increases and



the ability to handle alternating current is reduced. This is circumvented by using polycarbonate, but at higher cost. Paper dielectric capacitors are available in voltage ratings from 200 to 2000V d.c. and the values of capacitances available depends on voltage rating (in the region of 1 to 50 $\mu$ F). Polycarbonate types, rated at 600V d.c., have values from 1 to 20 $\mu$ F and metallized paper types, rated at 200V d.c., have values from 25 to 150 $\mu$ F. U.K. agents are Auriema Ltd, 23 King Street, London W.3.  
**WW 305 for further details**

### Cathode-ray display unit

A self contained c.r.t. display unit, the EV8000 from Electronic Visuals, has a 100 x 80mm flat-faced tube, regulated power supplies, balanced input amplifiers for vertical and horizontal deflections, and brightness (Z axis) control via either



a simple blanking amplifier or an optional wideband linear amplifier. The vertical amplifier has a bandwidth of d.c. to 10MHz and all inputs are compatible with t.t.l. and d.t.l. levels. The unit can be supplied in various case types and is suitable for 19in rack-mounting. Electronic Visuals Ltd, P.O. Box 16, Staines, Middx.  
**WW 304 for further details**

### High-voltage multiplier discs

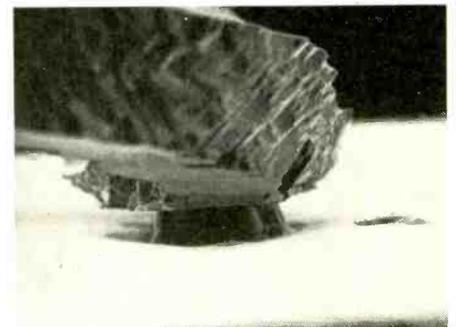
A series of 10kV high-voltage multiplier discs developed by Aerovox Corp and marketed by Auriema are made with high K dielectric materials. The coating is epoxy resin. Available in 1,000 and 2,000pF capacitances as standard (other values are available) the discs have a capacitance

tolerance of  $\pm 20\%$ , a 2% maximum dissipation factor, insulation resistance of 20,000M $\Omega$ , and are designed for working voltages up to 10,000V d.c. Other capabilities include a temperature characteristic of  $\pm 15\%$  change in capacitance from  $+10^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ , a corona voltage of approx. 2,700V r.m.s. at 50Hz and a corona level of less than 100 picocoulombs. Auriema Ltd, 23-31 King Street, London W.3.

**WW318 for further details**

### 60-GHz varactor diode

A gallium arsenide varactor diode capable of operating at 60GHz has been developed by the Services Electronics Research Laboratory and is manufactured by Marconi. The diode, of the diffused mesa type, is used in the 50-cm circular waveguide system of TE<sub>01</sub> mode propagation.



Known as type XMD3A, it is available in three figures of merit—from 40 to 90GHz—in a ceramic leadless inverted device package for simple integration into stripline structures. Photograph shows the gold connection to the chip magnified 2000 times. Marconi Co Ltd, Chelmsford, Essex.

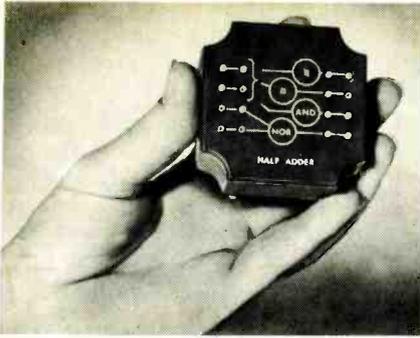
**WW 308 for further details**

### High-speed leadless diodes

Switching diodes for use in hybrid large- and medium-scale integrated circuits are made by Dickson Electronics Corp (U.S.A.). They are rated for 75mA forward current at 75V d.c. with a dissipation of 250mW. The diodes can be mounted upright with wire bonding or inverted with reflow solder bonds. Dual types are made with either common anode or common cathode connections. Available from Dage (G.B.) Ltd, Haywood House, 64 High Street, Pinner, Middx.  
**WW 310 for further details**

### Logic modules

Feedback Instruments have added a reed relay unit RU336 and a lamp display unit LD265 to their range of logic and analogue teaching elements. The RU336 contains two reed relays which, when either is operated by a logic '1' input, will switch currents up to 1A in external circuits, thus providing single-pole changeover switching controlled by the logic. The LD265 supplements the display lamps



already available on the Logikit mounting decks. Four buffered lamp circuits operate in the presence of a '1' at the input. All the logic (Logibit) elements are of identical size and are compatible with any of the Feedback logic teaching equipment. Prices of individual elements vary but all are under £10. Feedback Instruments Ltd, Park Road, Crowborough, Sussex.  
**WW316 for further details**

### Ferrite limiters

A series of solid-state ferrite limiters introduced by EMI-Varian is claimed to give a ten-fold increase in life expectancy over conventional TR tubes and limiters. The VFX 9500 device for example operates from 8.5 to 10.0GHz (X-band). The integral unit consists of a ferrite limiter followed by a diode limiter. The ferrite portion provides about 10dB of high-power isolation while the diode limiter reduces spike and flat leakage power to levels which provide reliable protection to the receiver diode.

Operating characteristics include:

bandwidth	any 5% b.w. from 8.5 to 10.0GHz
peak power	10kW
v.s.w.r.	1.5:1 max.
insertion loss	1.0dB max.
spike leakage energy	0.02 erg max.
flat leakage energy	20mW max.
recovery time	0.5µs max.
insertion length	88.9mm
weight	680g

EMI-Varian Ltd, Hayes, Middlesex.

**WW312 for further details**

### V.H.F. frequency synthesizer

A frequency synthesizer for 27 to 70MHz and tunable in increments of 25kHz is made by Akers Electronics, a Norwegian

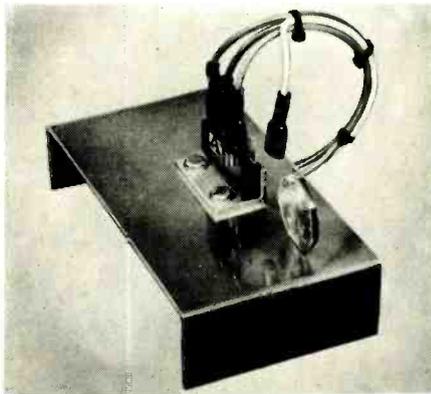


microelectronics manufacturing company. Using thin-film hybrid circuits and designed as a phase-locked loop it occupies only 86 × 60 × 40mm and weighs 0.35kg. It is intended as a frequency source in simplex military f.m. 'manpack' transceivers in the 27 to 70MHz band. During transmission it operates as a modulated exciter and during reception as local oscillator, when frequency range is 37.7 to 80.7MHz for an i.f. of 10.7MHz. Akers Electronics, 3191 Horten, Norway.

**WW 301 for further details**

### Power diodes in plastic

Two new power circuits from AEI handle double the power of earlier circuits. Type PM7A-Q, a bridge rectifier with 16-amp mean output and 44-watt dissipation, is available in seven voltage ratings from 200 to 1400 V r.m.s. Type PM6A-Q, two separate diodes of 10-amp mean current and 34 watts dissipation, is also available in the seven ratings. (The other kind of circuit PM5A is a diode-thyristor combination, but remains as originally



specified.) These three combinations are used in BDA 'Hotpoint' washing machines and enable universal motors to be used in place of expensive induction motors with the multiple windings required by electro-mechanical regulators. AEI Semiconductors Ltd, Carholme Road, Lincoln.

**WW 302 for further details**

### U.H.F. field-strength indicator

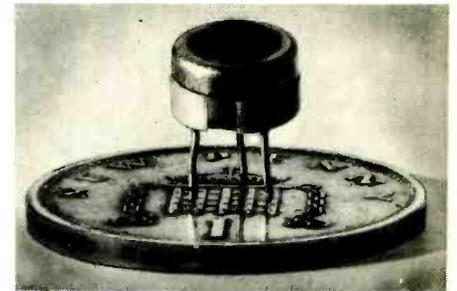
Designed and manufactured by Rohde & Schwarz the u.h.f. field-strength indicator, type Huze, comprises a transistor receiver and a log-periodic broadband aerial. The aerial is fixed to the receiver during measurement and can be adjusted in any direction and plane of polarization. The signal picked up is applied via a decade attenuator and a tunable bandpass filter to a mixer diode, where it is converted to the first i.f. of 150MHz. The instrument is operated from built-in sealed chargeable batteries. Battery power is sufficient for about eight hours with the a.f. section switched on (charging the storage batteries takes about 14 hours).

The charger may also be used as the power supply, the storage batteries acting as buffers. Field-strength range is 31 to 110dB above 1µV/m and voltage range 16 to 90dB above 1µV. Aveley Electric Ltd, Arisdale Avenue, South Ockendon, Essex.

**WW 307 for further details**

### 6-mm potentiometer

Sub-miniature metal glaze potentiometers to MIL specifications are made by TRW Inc. Available from 10Ω to 1MΩ in ½-watt rating (at 85°C) with temperature



coefficient of 1 in 10<sup>4</sup>/deg C. It measures 6.4mm dia and 4.5mm high. Type 171 is for vertical adjustment and type 172 for horizontal adjustment. Available in the U.K. from Dubilier Ltd, Victoria Road, London W.3.

**WW 306 for further details**

### I.C. for active filters

Three identical amplifiers are included on an integrated circuit made by Mullard. Designed for use in RC active filters up to 150kHz, each amplifier has an input resistance of 25kΩ and an output resistance of 9kΩ. Gain is 39dB or 117dB in cascade. An emitter follower is included which reduces output impedance to 500Ω. Designated type TAA 960 it operates from a 6-volt supply and consumes about 2mA. Mullard Ltd, Torrington Place, London, WC1E 7HD.

**WW 309 for further details**

### D.I.L. pulse transformer

Bourns (Trimpot) have introduced Model 4252-1005 miniature d.i.l. pulse transformer—a 16-pin unit with high insulation resistance, fast rise and fast fall time, clean pulse performance and low coupling capacitance. The specifications include:

operating temperature range	0° to 70°C
pulse inductance	150µH
(±10%, 0° to 70°C)	
leakage inductance	1.0µH
coupling capacitance	5pF
pulse width	400ns

Bourns (Trimpot) Ltd, Hodford House, 17/27 High Street, Hounslow, Middx.

**WW319 for further details**

# Literature Received

For further information on any item include the appropriate WW number on the reader reply card

## ACTIVE DEVICES

We have received the following literature from General Instrument Microelectronics, Stonefield Way, South Ruislip, Middlesex

'An introduction to the Glant family' (m.o.s., l.s.i.) ..... WW401

'Reliability aspects of low-voltage nitride' (m.o.s.) ..... WW402

A large number of data sheets on insulated-gate n-channel, p-channel, enhancement- and depletion-mode f.e.t.s ..... WW403

The English Electric Valve Co. Ltd, of Chelmsford, Essex, have published a 40-page reprint of a section of their main catalogue called 'Hydrogen thyratrons—preamble'. Principles of operation, terminology and circuitry are all discussed and a list of thyratrons available from E.E.V. is given ... WW404

Akers Electronics, 3191 Horten, Norway, have sent us the following data sheets:

AE501/2, positive/negative voltage regulator, < 30V, 0.6W, TO-5 ..... WW405

AE801/3, semiconductor transducer ..... WW407

AE830, miniature pressure transducer ..... WW408

AE61, semiconductor accelerometer ..... WW409

AE900 series, infrared detectors, visible light to 5.6 microns ..... WW410

S1401P, integrated active modulator ..... WW411

UH3000, thin film substrate containing 34 unconnected resistors from 100Ω to 10kΩ and four semiconductor chip mounting areas for constructing your own hybrid using a lead bonding machine ..... WW412

UH3011, integrated light sensitive Schmitt trigger ..... WW413

UH3027, integrated relay/lamp driver, 28V, 300mA ..... WW414

UH3021/3, analogue gate ..... WW415

A number of data sheets devoted to f.e.t.s ..... WW416

Nortronic A/S, 1380 Heggedal, Norway, produce the following data sheets:

Active filter unit (Butterworth) ..... WW417

Detector Unit (bridge), < 150kHz ..... WW418

Squaring unit, 80dB dynamic range ..... WW419

## PASSIVE COMPONENTS

'A designer's guide to battery systems' describes the construction, operation and use of mercury cells and gives an equivalents list. Mallory Batteries Ltd, Gatwick Rd, Crawley, Sussex ..... WW420

'Co-ordinated manual controls' is the title of a 16-page catalogue produced by the Microswitch Division of Honeywell Ltd, Windsor Rd, Slough, Bucks. It gives details on a matching range of control switches, push-buttons and indicator lamps. The square panel outline (56mm) of the devices is divided into four uniform display areas ..... WW421

## APPLICATION NOTES

A pamphlet lists accessories for audio equipment manufactured by the British Radio Corporation (Ferguson, H.M.V., Marconiphone, Ultra). The list includes slide synchronizing equipment, leads micro-

phones, head sets and the like. British Radio Corporation Ltd, P.O. Box 121, Lea Valley Trading Estate, Angel Rd, London, N.18 ..... WW422

Monograph # 3 'Real time signal processing in the frequency domain' is published by the Federal Scientific Corp., 615 West 131st St, N.Y.10027, U.S.A. .... WW423

RCA Ltd, Lincoln Way, Windmill Rd, Sunbury-on-Thames, Middlesex, have produced the following application notes:

AN4474 'Audio applications of the RCA HC1000 hybrid linear power amplifier' ..... WW424

AN4483 'General application considerations for the RCA HC1000 hybrid linear power amplifier' ..... WW425

AN4421 '16 and 25W broadband power amplifiers using RCA 2N5918, 2N5919 and TAA 7706 u.h.f. microwave power transistors' ..... WW426

A 30-page booklet 'A review of some main applications of Melinex polyester film' has been published by Industrial Film Sales Department, Imperial Chemical Industries Ltd, Plastics Division, P.O. Box 6, Bessemer Rd, Welwyn Garden City, Herts. .... WW427

General Instrument Microelectronics, Stonefield Way, South Ruislip, Middlesex, have published the following application notes:

T-6069, 'R.F. applications of the n-channel, dual-gate m.o.s.f.e.t. MEM554' ..... WW428

T-6032, 'A 30MHz amplifier stage utilizing a 2N4353' ..... WW429

T-6082, 'F.M. receiver using the n-channel, dual-gate, m.o.s.f.e.t. MEM554C' ..... WW430

T-6167, 'Uses of shift registers for data storage' ..... WW431

'Applying the m.o.s. read-only memory' ..... WW432

'F.E.T. application guide' (table) ..... WW433

## EQUIPMENT

A leaflet describes a low-cost (£56) d.c. to 4.5MHz oscilloscope with 5 × 4cm screen. The 'scope has a 100ns to 10ms calibrated timebase, an input of 1MΩ and 15pF, a sensitivity of 50mV/cm and X and Z inputs. Meteronic, Birchen Napps Platt, Nr. Sevenoaks, Kent ..... WW434

Reprints of the following articles which originally appeared in *Broadcast News* are available from R.C.A. Ltd, Lincoln Way, Windmill Rd, Sunbury-on-Thames, Middlesex:

'How to get the best pictures from the TK-27 film system' ..... WW435

'New a.m. ampliphase transmitters' ..... WW436

A family of timers covering the range 0.025 to 5min is described in a leaflet produced by Tempatron Ltd, 65 Milford Rd, Reading, Berks. .... WW437

Among the products manufactured by N H Research Inc., of California, is a precision a.c. source with a stability of ±10 parts in 10<sup>6</sup> per week. This, together with other products, is described in a booklet available from Lyons Instruments Ltd, Valley Works, Ware Rd, Hoddesdon, Herts. .... WW438

Aveley Electrics Ltd, South Ockendon, Essex, have sent us the following literature which originated from Rhode and Schwarz of West Germany and Scientific-Atlanta of Georgia, U.S.A.

Rhode and Schwarz:

'Semitest 111' (ISP) for testing digital integrated circuits. Can be used on r.t.l., d.t.l., d.t.l.z., t.t.l., e.c.l., and m.o.s. logic ..... WW439

'Power signal generator' (SLRE), 6.7 to 12.7GHz, 0.5 to 3W ..... WW440

'Modules for data acquisition and processing' (UC) ..... WW441

'Antenna rotators' (HA455/3 & 555/1) with hydraulic drive for directional aerial systems ..... WW442

'V.H.F. compact directional finder' (NP8), 117.5 to 136.5MHz ..... WW443

Scientific-Atlanta:

'Swept-frequency microwave measuring system' (series 1700) ..... WW444

'Horn antenna' (series 6800) ..... WW445

'Transportable u.h.f. telemetry tracking system' (series 3000-R18) 1435 to 1535 and 2200 to 2300Mhz ..... WW446

A new catalogue has been produced by Eagle International, Coptic St, London WC1A 1NR., which lists audio and test equipment as well as a variety of components ..... WW447

We have received the following literature from Lyons Instruments, Valley Works, Ware Rd, Hoddesdon, Herts., which describes goods manufactured in Switzerland by Institut Straumann.

OSC104, hermetically sealed tuning-fork oscillators operating at frequencies in the range 1 to 6kHz at temperatures from -55 to +85°C. Power supply between 5 and 12V ... WW448

DIV-, MUL-, FOS-104, frequency divider, frequency multiplier and sine-wave shaping unit for use with the tuning-fork oscillators WW449

STP-70/A, price list for the above range . WW450

A leaflet is available which describes the 'Miniscope' low-voltage soldering iron manufactured by Enthoven Solders Ltd, Dominion Buildings, South Place, London ECM 2RE ..... WW451

Nortronic A/S, 1380 Heggedal, Norway have sent the following literature:

703, multi-filter 35Hz to 14kHz. Contains 26

octave filters ..... WW452

Sound level indicator; versatile sound measuring set ..... WW453

Logic simulator for tuition, etc. .... WW454

702, universal filter 0.2Hz to 20kHz in 5 bands ..... WW455

701, universal filter 40Hz to 16Hz ..... WW456

Automatic tracking filter and wave analyzer (audio) ..... WW457

Intended for use with Tektronix 540 and 550 oscilloscopes a light coupled oscilloscope plug-in unit is described in literature available from Lyons Instruments Ltd, Valley Works, Ware Rd, Hoddesdon, Herts. Called the IsoAmp model 6150 the unit provides 1.5kV isolation and will accept signals up to 300V peak-to-peak at up to 35MHz. The signal is converted to light and the only connection to the oscilloscope is via a fibre optic light guide ..... WW458

The leaflet 'Short-wave puts you where it's at' describes the range of Hallicrafters communications receivers. The Hallicrafters Co, 600 Hicks Rd, Rolling Meadows, Illinois 60008, U.S.A. ... WW459

Second-hand computers are the subject of a catalogue from Computer Sales and Services, 49/53 Pancras Rd, London N.W.1. .... WW460

Kampel Electronics Ltd, 99 Old Christchurch Rd, Bournemouth, BH1 1EP, have produced a leaflet describing a stereophonic source simulator which is designed to be used with a stereo amplifier to produce a stereophonic effect from a monophonic signal source ..... WW461

## GENERAL INFORMATION

The Scientific Instrument Manufacturers' Association of Great Britain, SIMA House, 20 Peel St, London W.8, have produced a second edition of their 'Metrication Guide'. This revised edition costs £2.50

# April Meetings

*Tickets are required for some meetings: readers are advised, therefore, to communicate with the society concerned*

## LONDON

1st. IERE—"Future techniques for cockpit display" by D. R. Evans and Capt. D. S. Kirkland at 18.00 at 9 Bedford Sq., W.C.1.

1st. RTS—"Sound techniques for television and commercial recording" at 19.00 at Intersound Recording Studios, Park Drive, Wembley, Middlesex.

6th. IEE—"Business, society and the professional engineer" by B. M. Maskell at 17.30 at Savoy Place, W.C.2.

7th. IEE—"The use of telecommunications in meteorology" by A. A. Worthington at 17.30 at Savoy Place, W.C.2.

15th. IEE—Discussion on "Development of electrical and electronic systems during flight testing of new aircraft" at 18.00 at Savoy Place, W.C.2.

15th. RTS—"New techniques in video mixing" by W. R. Hawkins at 19.00 at I.T.A. 70 Brompton Rd., London S.W.3.

16th. IFF—Discussion on "Engineering science in schools and further education" at 14.00 at Savoy Pl., W.C.2.

16th. IEE—"The effective training of professional engineers as managers" by P. H. L. Thomas at 17.30 at Savoy Pl., W.C.2.

19th. IEETE—Panel under the chairmanship of Prof. R. C. G. Williams discussing "Panel connecting problems" at 18.00 at Savoy Pl., W.C.2.

20th. IERE—"Automatic camera line-up in colour television" by D. V. Ryley and Mrs. G. Claydon at 18.00 at the London School of Hygiene and Tropical Medicine, Keppel St., Gower St., W.C.1.

20th. AES—"Wide range ribbon loudspeaker development" by Stanley Kelly at 19.15 at the Mechanical Engineering Dept., Imperial College, Exhibition Rd., S.W.7.

21st. Inst. Navigation—"The application of low light television to navigation" at 17.00 at the Royal Geographical Society, 1 Kensington Gore, S.W.7.

21st. SERT—"System 24" by C. P. Davies at 19.00 at London School of Hygiene & Tropical Medicine, Keppel St., W.C.1.

21st. BKSTS—"Recording techniques for multichannel stereo" by Michael A. Gerzon at 19.30 at I.T.A., 70 Brompton Rd., S.W.3.

23rd. IEE—Discussion on "Light emitting diodes and their utilization" at 17.30 at Savoy Pl., W.C.2.

27th. IERE—"Communications through space, flame and fibre" by Prof. P. J. B. Clarricoats at 18.00 at the London School of Hygiene & Tropical Medicine, Keppel St., Gower St., W.C.1.

28th. IERE—"Applications of Camac" by H. Bisby at 18.00 at 9 Bedford Sq., W.C.1.

29th. IEE/IERE—Colloquium on "Design and application of minimal computers" at Savoy Pl., W.C.2.

29th. RTS—Fleming Memorial Lecture: "Perspectives in television" by Huw Wheldon at 19.00 at The Royal Institution, Albemarle St., W.1.

## BALLYMENA

20th. IERE—"Loudspeakers" by R. L. West at 19.30 at Ballymena Technical College.

## BIDEFORD

20th. British Computer Soc.—"Computer aided design" by B. Gott at 18.15 at the Library Theatre.

## BIRMINGHAM

19th. SERT—"Electronic music" by A. Douglas at 19.30 at Aston University.

21st. RTS—"Radio telescopes" by Dr. Guy Pooley at 19.00 at the A.T.V. Studio Centre, Bridge Street.

## BOURNEMOUTH

21st. SERT—"Remote control and indication systems" by N. Greene at 19.30 at Bournemouth Municipal Technical College.

## CHELMSFORD

28th. IEE—"Radio meteorology" by Dr. J. A. Saxton at 18.30 at King Edward VI Grammar School, Broomfield Rd.

## EVESHAM

20th. IERE—"World satellite communication systems" by G. H. Banner at 19.00 at B.B.C. Club.

## HIGH WYCOMBE

28th. IEE—"Plasma—the fourth state of matter" by Dr. J. E. Allen at 19.15 at the High Wycombe College of Technology and Art.

## LIVERPOOL

21st. IERE—"Computer aided circuit design" by E. Wolfendale at 19.00 at the Department of Electrical Engineering, University of Liverpool.

## MANCHESTER

15th. IERE—"Radio astronomy" by E. J. Daintree at 19.15 at the Renold Bldg., U.M.I.S.T., Altrincham St.

22nd. SERT—"Integrated circuits" by J. Tomson at 19.30 at U.M.I.S.T.

## NEWCASTLE-UPON-TYNE

14th. IERE—"Electronics and the entertainment industry—the future" at 18.00 at Ellison Building, The Polytechnic, Ellison Place.

## PLYMOUTH

7th. RTS—"Computers in television programming" by N. W. Green at 19.30 at the Studios of Westward Television Ltd.

## READING

29th. IERE—"Direct digital control—the case for the special purpose computer" by P. Atkinson and A. J. Allen at 19.30 at the J. J. Thomson Laboratory, University of Reading, Whiteknights Park.

## ROMFORD

21st. IERE—"Management of R & D" by D. C. Dalton at 18.30 at Central Library.

## ROTHERHAM

22nd. IEETE—"Yorkshire TV studios, Leeds" by P. G. Parker at 19.00 at College of Technology, Main Hall, Howard Street.

## SWANSEA

1st. IERE/IEE—"The use of satellites for civil communication" by C. F. Davidson at 18.15 at Department of Applied Science, University College.

## WALSALL

28th. IEETE—"Electronics in the automobile" by

W. F. Hill at 18.45 at Midlands Electricity Board District Offices, Green Lane.

## WOLVERHAMPTON

6th. IERE—"Direct digital control—the case for the small special purpose computer" by P. A. Atkinson at 19.30 at the Polytechnic.

## WORCESTER

22nd. B.Computer Soc.—"Introduction to computer simulation" by G. S. Perdue at 19.30 at Worcester Technical College.

# Conferences and Exhibitions

*Further details are obtainable from the addresses in parentheses*

## LONDON

Apr. 15 & 16 Imperial College  
Britain's Modern Standards—the contribution and the Gain

a (British Standards Institution, 2 Park St., London W1A 2BS)

Apr. 19 & 20 I.E.E., Savoy Place  
Hybrid Microelectronic Circuits

(International Society for Hybrid Microelectronics, c/o Dr. R. G. Loasby, A.W.R.E., Building A37, Aldermaston, Reading RG7/4PR)

Apr. 19-22 Alexandra Palace

Physics Exhibition  
(I.P.P.S., 47 Belgrave Sq., London S.W.1.)

Apr. 21-29 Earls Court

International Engineering and Marine Exhibition  
(Industrial & Trade Fair Ltd, Commonwealth House, New Oxford St., London WC1A 1PB)

## BRIGHTON

Apr. 4-6 University of Sussex

Vacuum Equipment  
(I.P.P.S., 47 Belgrave Sq., London S.W.1)

Apr. 20-23 Hotel Metropole

Technical Communication in the 70s  
(Business Conference & Exhibitions, Mercury House, Waterloo Rd., London S.E.1.)

## LANCASTER

Apr. 5-7 The University

Elementary Particle Physics  
(I.P.P.S., 47 Belgrave Sq., London S.W.1)

## YORK

Apr. 5-8 The University

Atomic and Molecular Physics  
(I.P.P.S., 47 Belgrave Sq., London S.W.1)

## OVERSEAS

Apr. 5 & 6 Atlanta

System Theory  
(C.O. Alford, School of Electrical Eng., Georgia Institute of Technology, Atlanta, Georgia 30332)

Apr. 12-15 Washington

Telemetry Conference  
(Washington Technical Consultants, 422 Washington Bldg, Washington D.C. 20005)

Apr. 13-15 Boston

Electronics in Medicine  
(Electronics in Medicine, 330 W. 42nd St., New York, NY 10036)

Apr. 13-15 New York

Computers and Automata  
(Polytechnic Institute of Brooklyn, 333 Jay St., Brooklyn, New York 11201)

Apr. 13-16 Denver

Magnetics Conference  
(C.D. Mee, IBM Corp., Building 015, Monterey & Cattle Rds, San Jose, California 95114)

# Real & Imaginary

by "Vector"

## Salute to "Free Grid"

Mention the name of Norman Preston Vincer-Minter to *W. W.* readers and I warrant that blank expressions will be the order of the day. But mention "Free Grid" and all of those of more than seven years' standing in readership will know at once who you're talking about. And will not only know, but almost invariably the mention of the name will provoke reminiscence of the "Do you remember what he said about—?" type.

But first, to the young among us, a word or two of explanation. The name N. P. Vincer-Minter appears in *W. W.* volumes prior to 1930, mostly as a contributor of constructional articles. Then, on September 17th, 1930, "Free Grid" made his entrance with a regular page feature under the heading "Unbiased". At first, to judge from his remarks, he was very much on trial (editors are adept at suspending the sword of Damocles) but before long his inimitable style made his place assured. For thirty-four years (he died, aged 67, on 11th March 1964) he continued to delight readers with his wit, his scholarship, his anecdotes and his "inventions". "Free Grid" became as much a part of the journal as the front cover.

At first the drawings which accompanied his page were relatively characterless but by the end of 1930 the "F.G." image (which, incidentally, bore no resemblance to Vincer-Minter) emerged, together with those of the formidable Mrs. "Free Grid" and the little "Grid Leaks". The illustrations became a perfect complement to the writing. Here, epitomized, was the middle-class little man of the period—bowler, spectacles, neat suit, furled umbrella and all—ever ready to do battle with bureaucracy, intransigent manufacturers, slipshod terminology or sheer stupidity. The aggressive chin of the cartoonist's little man added emphasis to the prose, for "F.G." was an arch-puncturer of balloons and the sworn enemy of sacred cows.

But it is in the role of prophet that "Free Grid" is best remembered. Here he was in a league of his own. In the Jubilee issue of *W. W.* in 1961 he really let himself go and as no commemorative issue would be complete without "Free Grid" I make no apology for turning over the rest of the page to him. Ladies and gentlemen (and particularly new readers), here is the incomparable "Free Grid" writing ten years ago:

### A.D. 1971, 1986, 2011

Another 50 years will have to pass before *Wireless World* can publish another jubilee number, and that will be the centenary number of April A.D. 2011. However, it is customary to celebrate 60th and 75th anniversaries of things. I shall be very surprised if by the 60th anniversary in 1971 we do not have coloured television and by the 75th anniversary in 1986 stereoscopic coloured TV.

By 1971 our television sets will probably have a scanning unit so that we can show our coloured slides and also our home ciné films on the c.r.t. and by 1986 our home ciné films will be returned to us from the processing station in the form of magnetic tapes holding both sound and vision recordings.

By 1986 every set will, of course, have a built-in multi-channel tape recorder for vision and sound so that while we are watching one programme we can simultaneously bottle one or more of the several alternative programmes that will be available.

### Fettered by Physics

I will now . . . venture to glance into the future of electronics but I am definitely not

going to inflict on you any of the unimaginative and rather obvious ideas which most science-fiction writers present to their readers. . . .

The reason for their unimaginative stories is that writers of science fiction allow their minds to be fettered by physics, or, more accurately, by our contemporary knowledge of physics. The "sciction" scribes, as I call them, write fantastic stories—doubtless accurate by contemporary scientific knowledge—about travel to distant worlds while overlooking the possibility of travel to another kind of world which is right under their noses. The world to which I refer is the extra-spatial and extra-temporal one which I discussed fully in the March, 1959, issue of this journal. I am greatly indebted to "Cathode Ray" for my ideas and gladly acknowledge it. As I explained in my original thesis on the subject it was he who set me thinking by his article in the November 1958 issue. In that article he gave us a very vivid picture of electrons as being "waves of which nobody knows" which it is usual to call  $\psi$  waves. As a result of reading this I expressed the view that if we could manage to alter one of the properties of the  $\psi$  waves, such, for instance, as their  $\lambda$ , we should probably

find that these metamorphosed electrons vanished, like H. G. Wells's Time Machine, out of our world of time and space into that extra-spatial and extra-temporal "world" inhabited by ghosts, fairies, poltergeists and other seemingly shadowy and clammy entities who seem to pass through brick walls, to be able to be in two places simultaneously and, in general, to ignore many if not all the laws of physics.

In actual fact I don't believe they do ignore them; they merely seem to ignore physical laws because our knowledge of physics today is very limited in comparison with what it will be in 2011.

I am reluctant to call this spaceless and timeless place the metaphysical world because I don't think it is "beyond physics" as the name would imply. I will, therefore, call it the psychotronic world which simply means that it is built of metamorphosed electrons, or, in other words, psychotrons, a word which I coined in the May 1960 issue to describe these extra-spatial and extra-temporal electrons or  $\psi$  waves which had had their wavelength or other property changed or metamorphosed and had, therefore become  $\mu\psi$  waves.

### Electrovision

I will venture only one prophecy on more ordinary lines . . . by suggesting that before 2011 our electronic experts and ophthalmic surgeons will have got together to do something very drastic for people like myself suffering from failing sight.

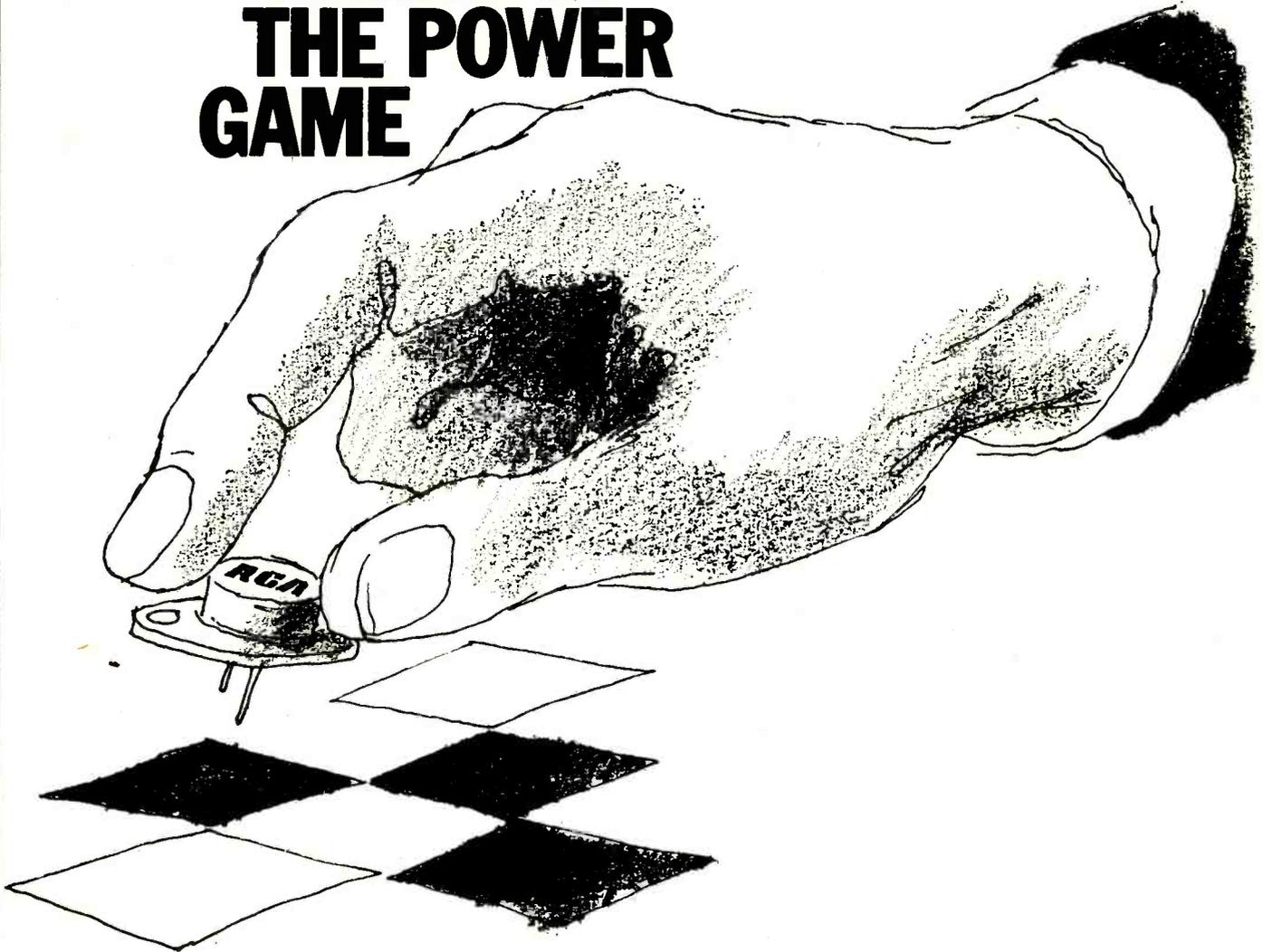
I have in mind the development of something like the special kind of cathode-ray tube used for transmission but in very miniature form so that it would actually take the place of an eye and convert vision into pulses along the optic nerve, as the natural eye does now.

*Requiescat in pace "Free Grid"*



"Books cannot always please, however good" was the caption to this cartoon illustrating "Free Grid" suffering from the symptoms of "uxorogenic cacophoria".

# NOW YOU CAN MOVE AHEAD IN THE POWER GAME



## RCA ANNOUNCE NEW POWER TRANSISTORS FOR QUASI COMPLEMENTARY AUDIO AMPLIFIERS.

Now available, six of the best from RCA.

Six new power transistors for output levels from 5W to 70W (8Ω impedance).

Manufactured by RCA to the highest professional standards.

Available from your local stockist as of now

Make a note of the right number for your project.

Type	40629	40630	40631	40632	40633	40636
Power Output	5W	7W	10/12W	25W	40W	70W

Or phone Sunbury-on-Thames 85511 and we'll tell you more. There's plenty more to tell.

**RCA Ltd., Solid State, Sunbury-on-Thames, Middlesex.**

OFFICIAL DISTRIBUTORS:

**Semicomps Northern Ltd.**, The Square, Kelso, Roxburghshire. Tel: 2366.

**REL Equipment & Components Ltd.**, Croft House, Bancroft, Hitchin, Herts. Tel: 50551/2/3 52202.

**ECS (Windsor) Ltd.**, Thames Avenue, Windsor, Berks. Tel. 68101 (20 lines).

# RCA Solid State

# Sinclair Project 60



the world's most advanced high fidelity modules

**Sinclair Project 60** presents high fidelity in such a way that it meets every requirement of performance, design, quality and value and now that the remarkable phase lock loop stereo FM tuner is available, it becomes the most versatile of high fidelity systems. With Project 60, it is possible to start with a

modest mono record reproducer and expand it to a sophisticated stereophonic radio and record reproducing system of fantastically good quality to hold its own with any other equipment, no matter how expensive. Project 60 is a unique high fidelity module system where compactness and ease of assembly are combined with

circuitry that is far in advance of any other manufacturer in the world. Thus it is extraordinarily easy to assemble any combination of modules using nothing more complicated than the simplest of tools, and you certainly do not have to be experienced to build with complete confidence. The 48 page manual free with Project 60 equipment makes everything easy and you can house your assembly in an existing cabinet, motor plinth, free standing cabinet or virtually any arrangement you wish. Once you have completed your assembly you will have superlatively good equipment to give you years of service and enjoyment. You will have obtained superb value for money because Project 60 is the best selling modular system in Europe and can therefore be produced at extremely competitive prices and with excellent quality control.

Sinclair Radionics Ltd., London Road, St. Ives, Huntingdonshire PE17 4HJ.  
Tel: St. Ives (048 06) 4311

**sinclair**

System	The Units to use	together with	Cost of Units
A Simple battery record player	<b>Z.30</b>	Crystal P.U., 12V battery volume control	<b>£4.48</b>
B Mains powered record player	<b>Z.30, PZ.5</b>	Crystal or ceramic P.U. volume control etc.	<b>£9.45</b>
C 20+20 W. R.M.S. stereo amplifier for most needs	<b>2 x Z.30s, Stereo 60, PZ.5</b>	Crystal, ceramic or mag. P.U., most dynamic speakers, F.M. tuner etc.	<b>£23.90</b>
D 20+20 W. R.M.S. stereo amplifier with high performance spkrs.	<b>2 x Z.30s, Stereo 60, PZ.6</b>	High quality ceramic or magnetic P.U., F.M. Tuner. Tape Deck, etc.	<b>£26.90</b>
E 40+40 W. R.M.S. deluxe stereo amplifier	<b>2 x Z.50s, Stereo 60 PZ.8, mains trsfmr</b>	As for D	<b>£34.88</b>
F Outdoor P.A. system	<b>Z.50</b>	Mic., up to 4 P.A. speakers controls, etc.	<b>£5.48</b>
G Indoor P.A.	<b>Z.50, PZ.8, mains transformer</b>	Mic., guitar, speakers, etc., controls	<b>£19.43</b>
H High pass and low pass filters	<b>A.F.U.</b>	C, D or E	<b>£5.98</b>
J Radio	<b>Stereo F.M. Tuner</b>	C, D or E	<b>£25.00</b>

# Sinclair Project 60

## Z.30 & Z.50 power amplifiers



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at full output and all lower outputs. Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

**SPECIFICATIONS (Z50 units are interchangeable with Z.30s in all applications).**

**Power Outputs**

**Z.30** 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.  
**Z.50** 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms, using 50 volts.

**Frequency response:** 30 to 300,000 Hz  $\pm 1$ dB.

**Distortion:** 0.02% into 8 ohms.

**Signal to noise ratio:** better than 70dB unweighted.

**Input sensitivity:** 250mV into 100 Kohms. For speakers from 3 to 15 ohms impedance.

Size  $3\frac{1}{2} \times 2\frac{1}{2} \times \frac{1}{2}$  in.

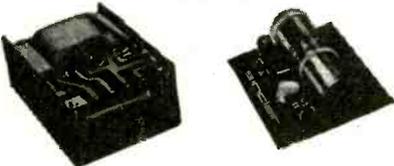
**Z.30**

Built, tested and guaranteed with circuits and instructions manual **£4.48**

**Z.50**

Built, tested and guaranteed with circuits and instructions manual. **£5.48**

## Power Supply Units



Designed specially for use with the Project 60 system of your choice.

Illustration shows PZ.5 to left and PZ.8 (for use with Z.50s) to the right. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

**PZ-5** 30 volts unstabilised **£4.98**

**PZ-6** 35 volts stabilised **£7.98**

**PZ-8** 45 volts stabilised

(less mains transformer) **£7.98**

**PZ-8** mains transformer **£5.98**

## Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.

## Stereo 60 pre-amp/control unit



Designed for the Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

**SPECIFICATIONS**

**Input sensitivities:** Radio—up to 3mV. Mag. p.u. 3mV; correct to R.I.A.A. curve  $\pm 1$ dB: 20 to 25,000 Hz. Ceramic p.u.—up to 3mV: Aux—up to 3mV.

**Output:** 250mV

**Signal-to-noise ratio:** better than 70dB.

**Channel matching:** within 1dB.

**Tone controls:** TREBLE + 15 to -15dB at 10KHz: BASS + 15 to -15dB at 100Hz.

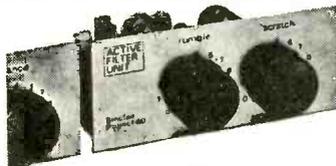
**Front panel:** brushed aluminium with black knobs and controls.

**Size:**  $8\frac{1}{2} \times 1\frac{1}{2} \times 4$  ins.

Built, tested and guaranteed.

**£9.98**

## Active Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two stages of filtering are incorporated—rumble (high pass) and scratch (low pass). Supply voltage - 15 to 35V. Current - 3mA. H.F. cut-off (-3dB) variable from 28k Hz to 5kHz. L.F. cut-off (-3dB) variable from 25Hz to 100Hz. Distortion at 1kHz (35V. supply) 0.02% at rated output.

Built, tested and guaranteed

**£5.98**

## Stereo FM Tuner



**first in the world to use the phase lock loop principle**

Before production of this tuner, the phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio over other systems. Now, for the first time, the principle has been applied to an FM tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Sensitivity is such that good reception becomes possible in difficult areas. Foreign stations can be tuned in suitable conditions and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.

**SPECIFICATIONS:**

**Number of transistors:** 16 plus 20 in I.C.

**Tuning range:** 87.5 to 108 MHz

**Capture ratio:** 1.5dB

**Sensitivity:** 2µV for 30dB quieting: 7µV for full limiting.

**Squelch level:** 20µV.

**A.F.C. range:**  $\pm 200$  KHz

**Signal to noise ratio:** >65dB

**Audio frequency response:** 10Hz—15KHz ( $\pm 1$ dB)

**Total harmonic distortion:** 0.15% for 30% modulation

**Stereo decoder operating level:** 2µV

**Pilot tone suppression:** 30dB

**Cross talk:** 40dB

**I.F. frequency:** 10.7 MHz

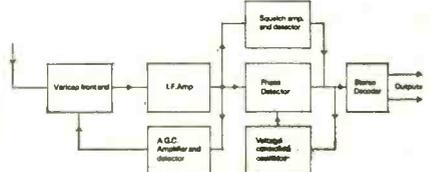
**Output voltage:** 2 x 150mV R.M.S.

**Aerial impedance:** 75 Ohms

**Indicators:** Mains on; Stereo on; tuning indicator

**Operating voltage:** 25-30 VDC

**Size:** 3.6 x 1.6 x 8.15 inches: 91.5 x 40 x 207 mm



Price: **£25** built and tested. Post free

To: SINCLAIR RADIONICS LTD LONDON ROAD ST. IVES HUNTINGDONSHIRE PE17 4HJ

Please send

\_\_\_\_\_

\_\_\_\_\_

\_\_\_\_\_

Name

\_\_\_\_\_

Address

\_\_\_\_\_

\_\_\_\_\_

for which I enclose cash/cheque/money order.

WW471

# Sinclair IC10/Q16/Micromatic

## IC10



### The world's most advanced high fidelity amplifier

This is the world's first monolithic integrated circuit high fidelity power amplifier and pre-amplifier. The circuit itself is a chip of silicon only a twentieth of an inch square by one hundredth of an inch thick, having 5 watts RMS output (10 watts peak). It contains 13 transistors (including two power types), 2 diodes, 1 zener diode and 18 resistors, and is encapsulated in a solid plastic package which holds the metal heat sink and connecting pins. This exciting device is more rugged and has considerable performance advantages, including complete freedom from thermal runaway and a very low level of distortion. The IC10 is primarily intended as a full performance high fidelity power and pre-amplifier, for which application it only requires the addition of such components as tone and volume controls and a battery or mains power supply. It may also be used in other applications including car radios, electronic organs, servo amplifiers (it is dc coupled throughout) etc.

### Circuit Description

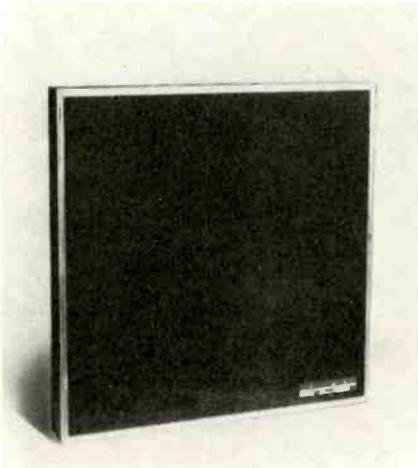
The first three transistors are used in the pre-amp and the remaining 10 in the power amplifier. Class AB output is used with closely controlled quiescent current which is independent of temperature. There is generous negative feedback round both sections and the amplifier is completely free from crossover distortion at all supply voltages, making battery operation eminently satisfactory.

Each IC10 is sold with a comprehensive manual giving circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include oscillators, etc. The pre-amp section can be used as an RF or IF, amplifier without any additional transistors.

### Specifications:

Output: 10 watts peak, 5 watts RMS continuous.  
 Frequency response: 5Hz to 100kHz  $1 \pm$  dB.  
 Total harmonic distortion: Less than 1% at full output.  
 Load impedance: 3 to 15 ohms.  
 Power gain: 110 dB (100,000,000,000 times) total.  
 Supply voltage: 8 to 18 volts. (A Sinclair power unit, PZ.7 is available for mains operation).  
 Size: 1 x 0.4 x 0.2 in. plus heat sink and tags.  
 Sensitivity 5 mV.  
 Input impedance: Adjustable externally up to 2.5 Mohms.  
 Price (with manual): £2.98 post free.

## Q16



### High fidelity loudspeaker

The Q16 employs the well proven acoustic principles specially developed by Sinclair in which a special driver assembly is meticulously matched to the characteristics of the uniquely designed cabinet. In reviewing this exclusive Sinclair design, technical journals have justly compared the Q16 with much more expensive loudspeakers. Its shape enables the Q16 to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures. A solid teak surround with a special all-over cellular foam front is used as much for appearance as its ability to pass all audio frequencies.

This elegantly designed shelf mounting speaker brings genuine high fidelity within reach of every music lover.

### Specifications:

Construction: Special sealed seamless sound or pressure chamber with internal baffle.  
 Loading: up to 14 watts TMS.  
 Input impedance: 8 ohms.  
 Frequency response: From 60 to 16,000 Hz, confirmed by independently plotted B and K curve.  
 Driver unit: Special high compliance unit having massive ceramic magnet of 11,000 gauss, aluminium speech coil and a special cone suspension for excellent transient response.  
 Size and styling:  $9\frac{1}{2}$  in square on face x  $4\frac{1}{2}$  in. deep with neat pedestal base. Black all-over cellular foam front with natural solid teak surround.  
 Price £8.98.

## Micromatic



### Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage AM receiver brilliantly designed to provide remarkable standards of selectivity, power and quality for its size. Powerful AGC counteracts fading from distant stations; bandspread at higher frequencies makes reception of Radio 1 easy. The plug-in magnetic earpiece provided matches the Micromatic's output to give wonderful standards of reproduction. Everything including the special ferrite rod aerial and batteries is contained within the minute and attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built and tested, you will find it as easy to take with you as your wrist watch, and dependable under the severest listening conditions.

### Specifications:

Size: 36 x 33 x 13 mm ( $1\frac{1}{5}$  x  $1\frac{3}{10}$  x  $\frac{1}{2}$  in.)  
 Weight: including batteries, 28.4 gm (1 oz.)  
 Case: Black plastic with anodised aluminium front panel and spun aluminium dial.  
 Tuning: medium wave band with bandspread at higher frequencies, (550 to 1,600 Hz).  
 Earpiece: Magnetic type.  
 On/off switching: By inserting and withdrawing earpiece plug.  
 Kit in pack with earpiece, case, instructions and solder £2.48.  
 Ready built, tested and guaranteed, with earpiece £2.98.  
 Two Mallory Mercury batteries type RM675 required. From radio shops, chemists, etc.

To: SINCLAIR RADIONICS LTD LONDON ROAD ST. IVES HUNTINGDONSHIRE PE17 4HJ

Please send

Name

Address

for which I enclose cash/cheque/money order.

VVW 4/71

Sinclair Radionics Ltd., London Road,  
 St. Ives, Huntingdonshire PE17 4HJ.  
 Tel: St. Ives (048 06) 4311

**sinclair**



# Contil INSTRUMENT CASES

Case No.	W	X	Y	Z
755	5	7	5	4
975	7	9	5	6
867	6	8	7	5
1277	7	12	7	6
10/27	12	16	7	11
1910/10	10	19	10	9
1910/100	10	19	10	9

(Panel size—10 1/2" x 19 x 4 1/2" with 5" vertical)

The ideal 'off the shelf' low cost instrument housing. 21-gauge steel. Finished hammer blue, with 18-gauge panel supplied with easy-to-strip protective covering for easy marking out. Individually packed, including feet and screws.

TYPE	10H	10off	100 off	P&P
755	£2.60	£2.45	£2.30	36p
867/975	£2.75	£2.65	£2.35	36p
1277	£3.10	£2.90	£2.70	45p
1277 un-	£2.50	£2.40	£2.05	45p
16127	£5.80	£5.60	£5.10	56p
1910/10	£7.90	£7.60	£7.25	66p
1910/100	£10.95	£10.80	£10.35	105p

## HANDLES

West Hyde carry stocks of four types of handles. The two Delrin types are 5.5/8" and 8 1/2" long with prices from 8p to 20p each. The two chrome panel handles are also in two sizes 3" and 5" with prices from 20p to 37p each. Both types of handles are ideal for our cases as the illustration shows.

## CHASSIS AND SPARE PANELS

Contil cases are also available with aluminium panels and Contilcore, applied after drilling and cutting. There is also a chassis to fit each size of Contil case. Three smaller sizes in 18 gauge aluminium and three larger sizes in 16 gauge. Prices from 15p to 75p.

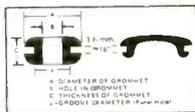
GROMMETS	500 off	PANEL HOLESIZES	100 off	500 off	BLANKING PLUGS
25p	20p	3/16"	35p	30p	
35p	30p	5/16"			
		3/8"			
		7/16"			
55p	50p	5/8"	55p	45p	
		1"	80p	70p	

Assorted £1.40 lots  
(100 3/16" 50 ea. of others) (100 off 3/16", 50 ea. of others)  
Postage & Packing 15p

## FEET

### You could tramp The World and not find better feet!

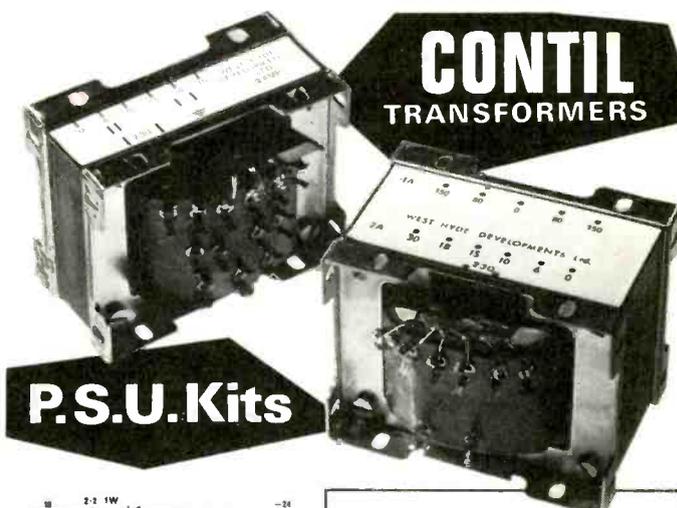
The West Hyde foot is moulded from a resilient, high hysteresis material, giving a high friction coefficient. Self-adhesive or screw fixing or both (6BA countersunk). Size 5/8" dia. 3/8" high. Grey. Price from 0.010 to 0.017 each.



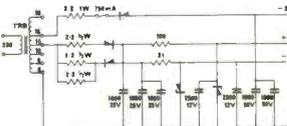
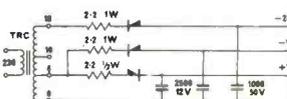
# WEST HYDE WH

WEST HYDE DEVELOPMENTS LTD., RYEFIELD CRESCENT, NORTHWOOD HILLS, MIDDX. HA6 1NN. Telephone: Northwood 24941/26732. Telex: 923231 Code: West Hyde Nthwd.

WW—104 FOR FURTHER DETAILS



## P.S.U. Kits



TRA has 100mA high voltage winding to drive neon tubes or neon indicators driven by transistors. Size: 4" x 3 1/2" x 3 1/2" O.P. 150-180 0-80-150 @ 100mA, 1, 6, 10, 15, 18, 30 @ 2 Amps. Therefore also gives 12-0-12 or 15-0-15 or 3, 4, 5, 6, 8, 10, 12, 18, 24 & 30 volts @ 2 Amps; and 300, 230, 180, 150, 80 and 70 volts @ 50/100mA. £2.50 P & P 37p.  
TRB. Size: 4" x 3 1/2" x 2 1/2". I.P. 230 volts. O.P. 0, 6, 10, 15, 18, 30 @ 2 Amps 1 T of TRA as above £1.95 P & P 30p.  
TAC. Size: 2 1/2" x 2 1/2" x 2 1/2". I.P. 230 volts O.P. 0-6-10-18 @ 1 Amp. Therefore gives also 4, 5, 6, 8, 10, 12, 18 volts. £1.45 P & P 22p.  
P.S.U. KITS  
P.S.U. KIT A. This kit contains a transformer, diodes, capacitors, resistors, tag strips, with chassis and neon indicator. £3.75 P & P 25p.  
P.S.U. KIT B. This power supply kit is based on the TRB transformer giving zener limited +V and -V supplies with a -24 volt line protected by a fuse. £7.40 P & P 30p.

All Contil Transformers, primarily intended for use with transistorised equipment, give a wide range of voltages.

# WEST HYDE WH

WEST HYDE DEVELOPMENTS LTD., RYEFIELD CRESCENT, NORTHWOOD HILLS, NORTHWOOD, MIDDX. HA6 1NN. TELEPHONE: NORTHWOOD 24941/26732. TELEX: 923231 CODE: WESTHYDE NTHWO

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# NEON CONTIL PUSH BUTTONS

## REED SWITCH

West Hyde reed switches have made 50,000,000,000 operations at over 2,000 per second. 1 off 0.75, 10 off 0.65 P & P 10p.

Already there are well over a million reed switches being used in Great Britain alone. The West Hyde version has the additional advantage of hermetically sealing and accurately positioning the reed in a protective polypropylene moulding.

**IDEAL FOR**  
OVER SPEED MONITORS, FLOW MONITORING, CONVEYOR MONITORING, COUNTING, POSITION DETECTION, PROXIMITY DETECTORS (Fe), REV. COUNTERS, LIMIT SWITCHES, PRESS TOOL PROTECTION.

**ITS ADVANTAGES ARE:**  
LOW COST, EASY MOUNTING, LONG LIFE, SAFETY, VIBRATION RESISTANCE, IDEAL FOR DRIVING LOGIC, HIGH OPERATING SPEED, HERMETICALLY SEALED CONTACTS.

Supplied with or without neon.		
1 or 2 microswitches.	} Same price	
Momentary or lock on.		
Red or natural colour.		
2 switches, illuminated		1 off 10 off
1 switch, illuminated		0.73 0.67
No switch, illuminated		0.63 0.55
1 switch, non-illuminated		0.44 0.39
2 switches, non-illuminated		0.43 0.38
		0.58 0.50
		less for quantities

**SPECIFY**  
1 or 2 microswitches  
Lock or momentary  
240 or 110v or no neon  
Whether cap is required



# WEST HYDE WH

WEST HYDE DEVELOPMENTS LTD., RYEFIELD CRESCENT, NORTHWOOD HILLS, NORTHWOOD, MIDDX. HA6 1NN. Telephone: 24941/26732. Telex: 923231 West Hyde Northwd.

WW—106 FOR FURTHER DETAILS



## LABEL-MATE

The Mini Label Mate is a new miniaturised labeller, that is small enough to slip into the pocket, but big enough to do the job. Made in durable shock-resistant plastic, it has a full view embossing wheel with a precise action, which gives fast finger tip embossing with clear, sharp impressions on 12 1/2 x 3/8" tape. Labeller in Blue, Green, Yellow, Red. Tape in Black, Red Blue, Green, Gold.

Labeller: £1. P & P 8p.  
Tape reels: 33p. P & P 5p.

West Hyde Dial Vernier Calipers are magnificent value and an enormous help on both development work and production. The direct reading dial is very clear, and they provide inside, outside, depth and step measurements. Satin chrome stainless steel with hardened slides. Packed in a velvet-lined metal case; available in four sizes: 6" x .001 or 150mm x .05mm @ £10.75. 12" x .001 or 300mm @ £22.75. Less for quantity.



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WEST HYDE DEVELOPMENTS LTD., RYEFIELD CRESCENT, NORTHWOOD HILLS, NORTHWOOD, MIDDX. HA6 1NN. Telephone: 24941/26732. Telex: 923231 WEST HYDE NTHWO.

WW—107 FOR FURTHER DETAILS

**OSCILLOSCOPE PROBE TM8119**  
High impedance 100/1 resistive attenuated probe for accurate display of HF waveforms or short rise time pulse signals, offered brand new with all accessories and instruction manual. List price £17. Our price £7.50 including earth bayonet TM8194. A MARCONI PRODUCT

**HIGH VOLTAGE TRANSFORMERS**  
Input 240 v., output 2560 v. and 2820 v. at 1 amp. Weight 75 lb. Price £15.

**AUDIO OSCILLATORS**  
TS 382/U

Range 0-200kHz in 4 ranges. Output voltage 1 micro volt to 12 volt, in seven ranges. Frequency check meter 60 and 400 Hz. Very good stability and low distortion. Contains thermostatically controlled heater. Supplied complete with leads circuit diagram etc. in as new condition. Price £35 P.P. £1.  
\* MANY OTHER TYPES AVAILABLE \*

**SOLARTRON OSCILLOSCOPE**  
523S.2

The best of the surplus scopes for £52, fully serviced and calibrated, compare the specification with others. Bandwidth DC-10MHz at 3 dB. Sensitivity is 1 MV/cm. Time Base 0.1 usec./cm/sec in 7 decades with fine control on each range. Uses C Core mains transformers/4 in. High resolution flat face PDA CRT and many other features make this scope very suitable for colour television servicing and many other applications. Price £52 P. & P. £1.25.

**BARGAIN OFFER 6V DC TAPE RECORDER MOTORS** Type DM148-1. Fully screened \* reversible \* constant speed \* specially designed for Portable Recorders \* Price only £1.75 P.P. 10p

**SCHOMANDL FREQUENCY METER TYPE FD.1 AND CONVERTER UNIT TYPE FDM.1**  
Range 1 KHz to 900 MHz an approved standard for telecommunications equipment. Offered calibrated to manufacturers specifications.

**CROYDON INSTRUMENTS**  
Precision Kelvin Wheatstone Bridge, type KWI. Measurements can be made from 0.0001 of an ohm, 100,000 ohms contains insitu Sullivan Galvo, four decade ranges, four standards and six Kelvin divide/multiply ratios offered in excellent condition ready for use. Price £95.

**MARCONI 12 KHz QUARTZ CRYSTAL** contained in BTG envelope with flying lead connections. Brand new only 62½p each.

**MORGANITE GLASS ENCLOSED RESISTORS** Value 2.5k. meg ohms, tolerance 10%. £1.25 per carton of four.

**WATSON MARLOW ORBITAL LOBE PUMPS**  
Specially designed for corrosive liquids etc. Rated output against 10 ft. head—110 G.P.H. direction of flow reversible. Supply 240 v. A.C. mains. Nett weight 14 lb. Supplied as new. Price £12.50. P. & P. 50p. List £22.50.

Voltage and Current regulators—heavy duty rheostats—1 ohm rated at 10A. Brand new by famous manufacturer, 62½p each. Also 1.5 ohm at 7A., 62½p, p.p. 7½p.

Lucas diode rectifiers—full wave bridge rectifier mounted on special heat-sink. 50V., 60V. operation rated at 50A. Has many uses for heavy duty charging plants, plating rectifiers, etc., etc. Per pair £8 (two complete bridge rectifiers), p.p. 37½p.

**GEC UNISELECTOR.** GPO pattern. 8 BANK 25 POSITION 75 ohm BRIDGING WIPERS. Brand new. Boxed. Only £2.50 P.P. 22p.

**RF SIGNAL GENERATORS AM AND FM**  
AVO Ltd. Model CT 378. Good quality AM generator 2-225 MHz in seven ranges—calibrated output level 1 uV to 1 V—frequency range directly calibrated with set level meter. Small size modern instrument complete with instructions. RF leads and mains lead for price only £35. Airmec Ltd. Model CT-212 AM/FM signal generator 85 kHz to 32 MHz directly calibrated output level calibrated 1 uV to 1 V deviation 0-30 kHz, fully portable for 24 DC and 240 v. AC operation in first class condition. Our price, only £45.



**SPECIAL OFFER**

**"INSULATION TESTERS" TYPE No. II METROHM** by famous British manufacturer. All solid state. No handles to crank. Runs off 9 volt transistor battery. Simply press button for function. Range 0.1 to 25M ohms for insulation testing. Also 0.1 to 100 ohms for resistance and continuity checking. Clear, concise scale. Small size modern instrument, complete with carrying strap and protecting cover. Offered in good used condition with battery ready to work. For 250 volt pressure only. List Price £19.50. Our Price £6.00 plus 22½p post/packing.

- Rhode & Schwarz ESM300 UHF Receiver AM/FM 85MHz—300 MHz.
- Rhode & Schwarz BN15031 Field strength test receiver AM/FM 90 MHz—470 MHz.
- Rhode & Schwarz BN4151/2\*60 Noise generator 3 MHz—1000 MHz.
- Rhode & Schwarz BN18042 Unbalanced standard Attenuator 0-100db 50 ohm 0 MHz—600 MHz.
- Rhode & Schwarz BN33664/50 UHF Load resistor 100 watt 50 ohm 0 MHz—600 MHz.
- Rhode & Schwarz BN4521 Vibration Meter 30Hz—12 KHz.
- Rhode & Schwarz ZD Diagraph.
- Advance Q meter type T.1. 100 kHz—100 MHz.
- Marconi Q meter type 329G 50 kHz—50 MHz.
- Marconi Q meter type 886A 15 MHz—170 MHz.

**DOUBLE BEAM OSCILLOSCOPE**  
SOLARTRON CD7115/2  
DC—7MHz CALIBRATED A1  
CONDITION, ONLY £65 P.P. £2

**RHODE & SCHWARZ POLYSKOP (SWOB 2)**  
With accessories for sale or hire.

Airmec portable RF signal generator. AM/FM Type CT212. Specially designed for field use for mains or 12v operation. Frequency range 85kHz to 30MHz. Accurate scale calibration. \*Variable output from 1 micro V 100mV 0 to 80db. Offered in excellent condition. Only £45.

**TEKTRONIX 551 WITH TWO PLUG INS PERFECT CONDITION**

**MARCONI 801D A.M. SIGNAL GENERATOR 10-470 MHz OUTPUT 0.1 uV to 1V**

Marconi TF887 Standard RF Signal Generator, range 15kHz to 30MHz. Variable output from 4 micro V to 4 Volts. Extremely accurate attenuator, high output stability and discrimination make the generator very suitable for precision measurements on networks and filters. Modulation up to 100% may be applied at 400 or 1000 Hz. Built in crystal calibrator. Offered in first class condition. Price £175.

Precision Multi Turn Indicating Dials suitable for 10 turn Helical Pots, machined from solid dural with the skirt engraved 0 to 100 and inner dial engraved 0 to 10 suitable for standard ¼ inch spindles, these small dials are as easy to fix as screwing on an instrument knob, size 1½ in. for skirt, 1½ in. dia. for counter knob depth ½ in. Brand new, only 77½p. A General Controls Manufacture.

**TEKTRONIX 581 WITH TYPE 80 PLUG IN AND PROBE AS NEW CONDITION.**

**WANTED. GOOD QUALITY TEST EQUIPMENT**

Miniature solenoid driven wafer switches, type-Ledex single pole, 7 pos., 3 wafers. Primarily used for channel switching in Radio-Telephones. Wafers may be substituted for any type. Solenoid voltage, 12 or 24V. Brand new. £1.50 each, p.p. 12½p.

CAMBRIDGE INSTRUMENT Co. Ltd. Precision test meters. Electrodynamic A.C. Ammeter 0 to 15 amps with test certificate £35  
Dynamometer A.C. Ammeter range 0 to 15 amps £45  
Cambridge Dynamometer A.C. test set 0-225 Watts/0-330 v./0-30 v. £55

Tinsley Universal Shunt type 4309C £5  
Tinsley Vernier Potentiometer type 4363E Auto £95  
Foster Thermocouple potentiometer type DX £75

Digital Voltmeter Solartron LM902-2 four digit readout £85  
Solartron A.C. Converter LM 903 matching unit for LM902 £75  
Hewlett Packard DVM 405CR four digit readout auto polarity £75  
Glouster DVM BIE 2123 A.C./D.C. transistor portable 0-1000 v. £75

**CANNON XLR AUDIO PLUGS AND SOCKETS**  
3 POLE and 6 POLE  
AVAILABLE EX STOCK  
BRAND NEW

**MARCONI 1094 A/S HF SPECTRUM ANALYSER**  
3-30MHz  
LATE MODEL FOR SALE OR HIRE

**SOLARTRON VF252/NSL PRECISION AC MILLIVOLT METER**  
Range 1.5 milli volt (for full scale deflection) to 15 volts in eight ranges input impedance 30 M ohms. The meters offered are of the very latest type not to be confused with the older models. Price only £75.

**LUCAS CAR RELAYS.** 12 v. Heavy duty make. Suitable for spotlights, horns, overdrives, etc. Brand new. Only 37½p. Special price for quantities.

**BARGAIN OFFER**  
200-yard reels equipment wire, size 1/024, STC quality, various colours. Brand new reels only 75p. P. & P. 12½p.

**LOW VOLTAGE POWER SUPPLY UNITS**  
To supply 12-15-20-24 and 30 volts- at continuous 5 amps with current control and ammeter employs silicon heavy duty rectification and high quality components very suitable for light duty plating and charging duties. 240 v. AC supply, fully fused. Small size only 10x7x6 in. Offered brand new units. Price £12.50.

**HUNTER MAGSLIPS** 3 inch Series, Type E-18-V.2. Very suitable for servo operation of hydraulic valves radar aerials and other applications for 50 volt 50 cycle operation. Offered brand new in transit boxes, at only £3.25 each.

**MUIRHEAD PHASEMETER**  
D-729-Im. Complete with supply and D925A Tunable Filter. Offered as new, with manual. Price £275.

**ADVANCE DC STABILIZED P.S.U. TYPE PM8**  
Fully stabilized power module PM8  
15 to 30 volts 5 amps offered brand new, Price £25

**50 DECO IMPULSE COUNTERS**  
4 DIGIT RESET  
10 Impulses per second.  
27MA 220V COIL AC/DC  
OFFERED BRAND NEW  
AT £2 EACH

**EIMAC SK-600A.** Air spaced Valve Holders suitable for 4X250, etc. Power tetrodes, brand new, boxed, complete with clamps, screws; heavy silver plate finish. Normal list price £6.50. Our price £2.50.

**A.E.I. MINIATURE UNISELECTOR SWITCHES**  
No waiting, straight off the shelf and into your equipment, the Catalogue Nos. are 2202A, 4/33A63/1; coil resistance is 250 ohms. Complete with base, and the price is £5. Limited quantity only available.  
Also: 2203A, 2200A, 2202A.

Resolved Components Indicator VP 253/1a. Solartron Low Frequency Decade Oscillators. Solartron OS 103 and associated equipment. 2 Phase Low Frequency Oscillator, type B0 567. Solartron. Solartron Synchro test set, type CT 428. Solartron AC Millivolt meter. Precision. Type VF 252.

**AERIAL CHANGE/OVER RELAYS** of current manufacture designed especially for mobile equipments, coil voltage 12 v., frequency up to 250 MHz at 50 watts. Small size only, 2 in. x ½ in. Offered brand new, boxed. Price £1.50, inc. P.&P.

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**COAXIAL SWITCHES**  
American Manufacture  
Suitable for aerial changeover and high frequency switching up to 1,000 MHz miniature Vacuum drawn type 110 vdc operation connections BNC and N types. Offered brand new, boxed. Price £3.25.

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Transparent casing. Size 2½x5x7 in. Offered brand new and boxed, 2 batteries per box, complete with links and full instructions. Can supply voltages in the range from 2-20v. Price £2.25, inc. P.&P.

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500 PIV Max rect. Current 16 amps. Guaranteed perfect. Price £1.25 each.

**COLVERN HELICAL POTS**  
1K ohms }  
5K ohms } ALL TEN TURN  
10K ohms }  
20K ohms }  
30K ohms }  
PRICE £1.75

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Capacitance range 0 to 100 pf fully screened with engraved vernier subdivided into 100 equal divisions complete with vernier index and original manufacturers seal offered brand new, at only £25 each.

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LOTS OF 100,000 — £150

10,000 — £20

1,000 — £3

500 — £2

## 1,000,000 GERMANIUM TRANSISTORS

LOTS OF 100,000 — £250

10,000 — £30

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**SPECIAL 50p PACKS. ORDER 10 PACKS AND WE WILL INCLUDE AN EXTRA ONE FREE!!!!**

RESISTORS, 1/2 watt assorted	100 50p	TRANSISTORS	
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SLIDERS	15 50p	Solid Core, Insulated	100yds. 50p
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NUTS AND BOLTS. Mixed length/type		Large Selenium	2 50p
8 B.A.	100 50p	Small	3 50p
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7 1/2in. x 3 1/2in.	6 50p	MIKES	1 50p
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**SINCLAIR AMPLIFIERS AND SPEAKERS:** Complete range in stock. All at 10% discount on list.

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2 1/2in x 1in x 0.15in	6p	5in x 3 1/2in x 0.15in	28p	3 1/2in x 3 1/2in x 0.1in	24p
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3 1/2in x 3 1/2in x 0.15in	20p	17in x 3 1/2in x 0.15in	74p	5in x 3 1/2in x 0.1in	28p
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Spot Face Cutter 38p. Pin Insert Tool 48p. Terminal Pins (0.1 or 0.15) 36 for 18p. Special Offer Pack consisting of 5 2 1/2in x 1in boards and a Spot Face Cutter—50p.

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**TRANSISTORISED FLUORESCENT LIGHTS, 12 volt.** All with reverse polarity protection. 8 watt type with reflector, suitable for tents, etc., £3. Postage/Packing 25p. 15 watt type, batten fitting for caravans £4. Postage/Packing 25p. 13 watt type, batten with switch, 22in x 2in x 1in £5. Postage/Packing 25p. THESE CAN BE SENT ON APPROVAL AGAINST FULL PAYMENT.

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1,000pf, 1,200pf, 1,500pf, 1,800pf, 2,200pf, 15p per dozen (all 400V working). 0.15uf, 0.22uf, 0.27uf, 30p per dozen (all 160V working). 25x discount for lots of 100 of any one type.

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1/2 and 1 watt. Most values in stock. 50p per 100. 10p per dozen of any one value. WIRE WOUND MAINS DROPPERS. Hundreds of values from 0.7 ohm upwards. 1 watt to 50 watts. A large percentage of these are multi-tapped droppers for radio/television. Owing to the huge variety these can only be offered "assorted" at 50p per dozen.

**SILVER MICA/CERAMIC/POLYSTYRENE CONDENSERS**

Large range in stock, 75p per 100 of any one value, 15p per dozen.

**RECORDING TAPE BARGAIN!** The very best British Made low-noise high-quality Tape! 5in Standard 38p. Long-play 45p. 5 1/2in Standard 45p. Long-play 60p. 7in Standard 60p. Long-play 82p. We are getting a fantastic number of repeat orders for this tape. Might we suggest that you order now whilst we still have a good stock at these low prices?

**£5 WORTH OF COMPONENTS FREE!!!!**

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In our opinion these units are the best value for money ever offered. A complete stereo unit consisting of Control unit, Pre-amplifier, Two Main Amplifiers and Power Pack complete the ready for use—NO extra components to buy—yours for £15. (Normal retail price is £16.50.)

**GARRARD SP 25 UNITS** also offered at a discount. Our price only £12.50. Postage 50p.

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Once again we have a supply of these excellent radios which offer superb quality sound and excellent sensitivity. They are packed in a colourful presentation box complete with battery, earpiece and carrying case. Each one is guaranteed. You would expect to pay at least £5—but our price due to bulk purchase is only £1.88.

# TRANSISTORS

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2N708A	12p	BC113	25p	BSY95A	15p
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2N1131	30p	BC115	32p	BY126	15p
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2N1306	25p	BC135	30p	BYZ15	£1.00
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2N2904A	32p	BCY32	50p	MPP102	42p
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2N2906A	32p	BCY38	40p	MPP105	40p
2N2907	37p	BCY39	60p	NKT212	25p
2N2926	12p	BCY40	50p	NKT214	15p
2N3011	25p	BCY41	15p	NKT216	37p
2N3053	25p	BCY42	15p	NKT217	40p
2N3054	50p	BCY43	20p	NKT277	20p
2N3055	75p	BCY58	25p	NKT403	75p
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2N3702	£1.10	BCY70	20p	OA5	20p
2N3702	12p	BCY72	15p	OA9	20p
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2N3705	15p	BCY79	30p	OA20	10p
2N3707	15p	BCZ10	35p	OA47	10p
2N3709	12p	BCZ11	40p	OA70	10p
2N3710	12p	BD112	50p	OA73	10p
2N3819	35p	BD121	65p	OA79	10p
2N3820	60p	BD123	80p	OA81	10p
2N4058	17p	BD125	50p	OA82	10p
2N4061	15p	BD131	75p	OA90	10p
2N4558	37p	BD132	85p	OA91	7p
2N4559	50p	BD153	62p	OA95	7p
28301	50p	BD156	57p	OA200	7p
28302	50p	BD181	75p	OA202	10p
28303	60p	BDY101	£1.25	OC16	50p
28304	75p	BDY11	£1.62	OC20	97p
40250	50p	BDY17	£1.50	OC22	50p
40361	55p	BDY18	£1.50	OC23	60p
40362	60p	BDY19	£1.75	OC24	60p
AA340	10p	BDY61	£1.97	OC25	37p
AA342	12p	BDY62	£1.97	OC28	62p
AAZ17	10p	BF115	25p	OC29	62p
AAZ17	10p	BF152	30p	OC35	50p
AC107	37p	BF154	40p	OC36	62p
AC126	25p	BF158	30p	OC41	25p
AC127	25p	BF159	60p	OC42	30p
AC128	25p	BF177	30p	OC43	40p
AC176	25p	BF178	25p	OC44	17p
AC187	30p	BF179	40p	OC45	15p
AC188	30p	BF180	37p	OC49	20p
AC197	30p	BF181	37p	OC71	15p
ACX18	25p	BF182	37p	OC72	25p
ACX19	25p	BF184	25p	OC73	30p
ACX20	22p	BF185	17p	OC74	30p
ACX21	22p	BF197	15p	OC75	25p
ACX22	17p	BF200	37p	OC76	25p
ACX39	50p	BF274	37p	OC77	40p
AD140	50p	BFW87	25p	OC81	25p
AD149	50p	BFW88	25p	OC82	25p
AD161	37p	BFW89	20p	OC83	25p
AD162	37p	BFW90	22p	OC84	25p
AF114	25p	BFY91	20p	OC139	25p
AF115	25p	BFY93	25p	OC170	62p
AF116	25p	BFY95	15p	OC171	62p
AF117	25p	BFY96	15p	OC172	30p
AF118	62p	BFY97	15p	OC173	30p
AF124	25p	BFY98	15p	OC200	40p
AF125	20p	BFY99	15p	OC201	60p
AF126	17p	BFY99	15p	OC202	75p
AF127	17p	BFY99	15p	OC203	40p
AF139	47p	BFY99	15p	OC204	40p
AF180	52p	BFY99	15p	OC205	75p
AF181	42p	BFY99	15p	OC206	90p
AF186	40p	BFY99	15p	OC207	90p
AF239	42p	BFY99	15p	OC271	97p
ASX26	25p	BFY99	15p	ORP12	50p
ASX27	32p	BFY99	15p	ORP61	42p
ASX28	25p	BFY99	15p	ZTX107	15p
ASX29	30p	BFY99	15p	ZTX300	12p
ASX67	40p	BFY99	15p	ZTX500	20p
ASZ21	42p	BFY99	15p	ZTX509	30p
BA115	7p	BFY99	15p	ZTX531	30p
BA164	10p	BFY99	15p	Discounts	
BAX13	6p	BFY99	15p	10% 12+	
BAX16	7p	BFY99	15p	15% 25+	
BAX17	7p	BFY99	15p	20% 100+	
BAX31	7p	BFY99	15p	Any one type	

# HENRY'S LOW COST INTEGRATED CIRCUITS

WE OFFER FROM STOCK AN EXCLUSIVE RANGE OF BRAND NEW CERAMIC FULL SPECIFICATION LOW COST TTL 7400 RANGE OF INTEGRATED CIRCUITS

Part No.	Description	Price 1-49	Price 50-99	Price 100-499	Price 500+
7400	Quad 2-Input NAND Gate	25p	20p	18p	15p
7401	Quad 2-Input NAND Gate Open Collector	25p	20p	18p	15p
7402	Quad 2-Input Positive Nor Gate	25p	20p	18p	15p
7404	Hex Inverter	25p	20p	18p	15p
7405	Hex Inverter with Open Collector	25p	20p	18p	15p
7410	Triple 3-Input NAND Gate	25p	20p	18p	15p
7430	Single 8-Input NAND Gate	25p	20p	18p	15p
7440	Dual 4-Input Buffer Gate	25p	20p	18p	15p
7441	BCD to Decimal Decoder and NIX Driver	£1.00	90p	80p	75p
7442	BCD to Decimal Decoder (TTL)	£1.00	90p	80p	75p
7453	Dual 2-Input and/or not Gate—Expandable	25p	20p	18p	15p
7453	Single 8-Input and/or not Gate—Expandable	25p	20p	18p	15p
7460	Dual 4-Input—Expandable	25p	20p	18p	15p
7470	Single JK Flip Flop—Edge Triggered	40p	35p	30p	25p
7472	Single Master Slave JK Flip Flop	40p	35p	30p	25p
7473	Dual Master Slave JK Flip Flop	45p	40p	35p	30p
7474	Dual D Flip Flop	45p	40p	35p	30p
7475	Quad Bistable Latch	50p	45p	40p	35p
7476	Dual Master Slave Flip Flop with Preset	50p	45p	40p	35p
7483	Four Bit Binary Counter	£1.00	90p	80p	75p
7490	BCD Decade Counter	£1.00	90p	80p	75p
7492	Divide by 12. 4 Bit Binary Counter	£1.00	90p	80p	75p
7493	Divide by 16. 4 Bit Binary Counter	£1.00	90p	80p	75p
7494	Dual Entry 4 Bit Shift Register	£1.00	90p	80p	75p
7495	4 Bit Up Down Shift Register	£1.00	90p	80p	75p
7496	5 Bit Shift Register	£1.00	90p	80p	75p

Data available for above series in booklet form, price 10p. Larger quantity prices Extn. 4 Dual Inline 14 Pin Sockets 30p each. 16 Pin 35p each.

# TRIACS GENERAL ELECTRIC

Type	P.I.	Cur. Volts	(All stud mounting)	1-49	50+	100+	500+
SC35A	100	3 amps	90p	75p	65p	60p	
SC35B	200	3 amps	95p	80p	70p	65p	
SC35D	400	3 amps	£1.00	85p	75p	70p	
SC40A	100	6 amps	£1.00	85p	75p	70p	
SC40B	200	6 amps	£1.20	£1.00	85p	80p	
SC40D	400	6 amps	£1.25	£1.10	£1.00	90p	
SC45A	100	10 amps	£1.25	£1.10	£1.00	90p	
SC45B	200	10 amps	£1.35	£1.20	£1.10	£1.00	
SC45D	400	10 amps	£1.50	£1.35	£1.20	£1.10	
SC50A	100	15 amps	£1.65	£1.50	£1.35	£1.20	
SC50B	200	15 amps	£1.75	£1.60	£1.45	£1.30	
SC50D	400	15 amps	£2.00	£1.75	£1.60	£1.40	
SC40E	500	6 amps	£1.50	£1.25	£1.10	£1.00	
SC45E	500	10 amps	£1.75	£1.50	£1.35	£1.25	
SC50E	500	15 amps	£2.25	£2.00	£1.75	£1.55	
DIAC	ST2	20p					

Larger quantity prices on application Extn. 4

# SILICON RECTIFIERS

Type	P.I.V.	1-49	50+	100+	500+	1000+
IN4001	50	8p	7p	6p	5p	4p
IN4002	100	9p	8p	7p	5p	4p
IN4003	200	10p	9p	7p	6p	5p
IN4004	400	10p	9p	8p	7p	6p
IN4005	600	12p	10p	9p	7p	6p
IN4006	800	15p	14p	12p	11p	9p
IN4007	1000	20p	16p	13p	12p	10p

Type	P.I.V.	1-49	50+	100+	500+	1000+
PL4001	50	10p	9p	8p	7p	6p
PL4002	100	11p	10p	9p	8p	7p
PL4003	200	12p	11p	10p	9p	8p
PL4004	400	12p	11p	10p	9p	8p
PL4005	600	15p	13p	11p	10p	9p
PL4006	800	17p	15p	13p	12p	10p
PL4007	1000	20p	17p	15p	13p	11p

Type	P.I.V.	1-49	50+	100+	500+	1000+
PL7001	50	20p	18p	17p	16p	14p
PL7002	100	20p	19p	18p	17p	15p
PL7003	200	22p	20p	19p	18p	16p
PL7004	400	25p	23p	21p	20p	18p
PL7005	600	26p	24p	23p	22p	20p
PL7006	800	27p	25p	24p	23p	21p
PL7007	1000	30p	28p	26p	24p	22p

Type	P.I.V.	1-49	50+	100+	500+	1000+
1002	100	2 amps	60p	55p	50p	45p
2002	200	2 amps	70p	65p	60p	55p
4002	400	2 amps	80p	75p	70p	65p
6002	600	2 amps	90p	80p	75p	70p
1004	100	4 amps	70p	60p	55p	50p
2004	200	4 amps	75p	70p	65p	60p
4004	400	4 amps	80p	75p	70p	65p
6004	600	4 amps	90p	80p	75p	70p
1006	100	6 amps	75p	70p	65p	60p
2006	200	6 amps	80p	75p	70p	65p
4006	400	6 amps	£1.10	£1.00	90p	80p
6006	600	6 amps	£1.25	£1.10	£1.00	90p

# R.C.A. INTEGRATED CIRCUITS

Linear Types	Price	Price	Price	Price
CA3005	£1.20	CA3035	£1.25	
CA3011	75p	CA3036	90p	
CA3012	90p	CA3039	85p	
CA3013	£1.10	CA3041	£1.10	
CA3014	£1.45	CA3042	£1.10	
CA3018	£1.40			

# HENRY'S RADIO LIMITED

ENGLAND'S LEADING ELECTRONIC CENTRES

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## HENELEC SELF-POWERED PRE-AMPLIFIERS



IC STEREO



FET 154



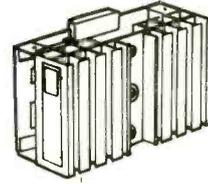
FET 9/4

Brochure No. 25 on request

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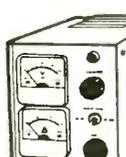
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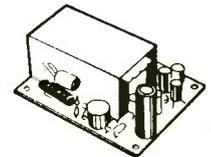
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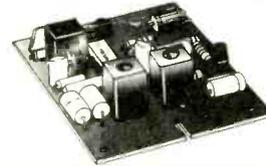
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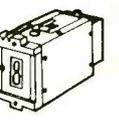
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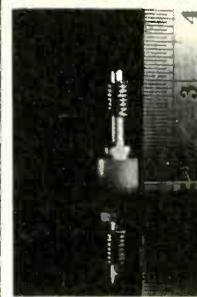
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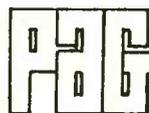
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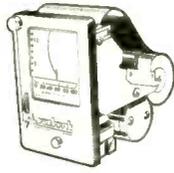
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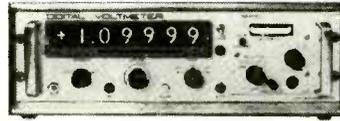


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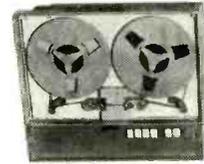
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Sedeco	ATCEZ4E	4	10	E/12V D.C.	12V D.C.	120mA	4"L x 2½" x 1½"	C.5	New Used	5.25 1.50
Sedeco	ATCEZ5E	5	25	E/24V D.C.	24V D.C.	240mA	4"L x 1½" x 2½"		New	6.00
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Counting Instrument	150G	4	15		24V D.C.			C.3	Each digit independently set, counts down to zero operating main switch	6.50
"	429	4	15	E/240V 50Hz	24V D.C.			C.12		4.12½
"	120	6	15	E/24V D.C.	24V D.C.		3½" L x 3½" x 1"			4.75
"	101A	6		M.	48V D.C.		4"W x 2½" x 2½"		Used	3.12½
Veeder Root	BD134545	5							Mechanical operation. Ratchet reset Inverse Nos.	0.82½
"	"	6		M.	160V D.C.					2.75
"	B38	6		M.	48V D.C.					2.75
"	"	6		M.	110V D.C.					2.00
"	"	6		M.	230V 50Hz					2.75
"	"	6		M.	24V D.C.					2.00
Haztler	"	6		M.	24V D.C.				500 Ω coll. New	4.50
"	"	6		M/E 110V D.C.	110V D.C.				1100 Ω/800 Ω. Used	2.45

Many other types of counters are available ranging from 3-6 digit with various supply voltages. Ring our Sales Office for further information.

TEKTRONIX Plug in Unit Type E—BRAND NEW. Price £75 P. & P. 50p. Also Type 80 £25

**BRAND NEW ELECTRO-MAGNETIC COUNTER**

A high precision counter offered at a fractional cost of other manufacturers of similar type. High counting speed. 25 impulse/sec. 6 digit display. 24 volt D.C. supply. 2.75 watts. 840 ohms. Size: 100mm x 50mm x 26mm. Immediate delivery. £4.50. (Carriage extra.) Other various voltage and impulse rates available. Phone or write for details.



NUMERICUS End Reading	Quantity	Price Each (Less Base)	Price Base
GR10M/U (Clear)	1-3	(£1.40) 28/-	Bases (20p)
	4-10	(£1.35) 27/-	4/-
	11-25	(£1.30) 26/-	Each
	26-100	(£1.20) 24/-	Less Bases
Side Reading			
XN8/FA 38 m/m lead (Amber)			
XN3/F 38 m/m lead (Red)	1-3	(£1.15) 23/-	
XN3A/F 6 m/m lead (Red)	4-10	(£1.10) 22/-	
XN3A 6 m/m lead (Clear)	11-25	(£1.05) 21/-	
XN11/F 38 m/m lead (Red)	26-100	(£0.95) 19/-	
XN23/FA 38 m/m lead (Amber)			Post Free

**EICHNER 8 HOLE PUNCH** No motor drive required. Solenoid operated equipment using 48V Reader £28.50; Punch £49.50. Carriage £1.25.

**7 HOLE NON PARITY TAPE PUNCH** New condition.

**LOW SPEED 7 HOLE TAPE PUNCH** 60 characters per second by well-known manufacturer.

**TELETYPE 8 HOLE PAPER PUNCH BRPE11 £260.** Also available 5 hole punch BRPE2 as above. This model has interchangeable heads. Complete with spooler. Price £75.

**5/7 HOLE OPTICAL READER BY FERRANTI** 20 characters per second. £20.

(183) SIGNAL GENERATOR CT 450 SANDERS. Range 7 KHz-12 KHz. O/p. 0-±50V. Attenuation range -10 to +100 dB. Price £85

**TRANSDUCER OSCILLATOR-AMPLIFIER-DEMOMULATOR.** An encapsulated unit for matching with S.E. Transducers. Suitable where space or adverse environmental conditions prevail. Supplied with a matching transducer a typical o/p is ± 3V into 50K Ohms. Supply voltage 12V. D.C. Range of transducers available 0.50-0.750; 0.1000; 0.4000 psi. Price £85

**TRANSDUCER—New Resistive Borden Tube Principle** pressure Transducer by K.D. Instrument. Model TD 216 0-2000 psi. Ref. C. 6. Price £15  
**TRANSDUCER NEW EX-GOVERNMENT DISPLACEMENT BONDED RESISTANCE STRAIN GAUGES.** Range ± ½ mechanical displacement equivalent to 0.3% resistive change. 3.5 + 3.5 Kohms. Model IT-2-31-35. Price £10

**OSCILLATOR.** High discrimination, by Marconi T.F. 1168. This instrument suitable for H.F. Communication. Due to its high discrimination makes it suitable for crystal filter response in Tx and Rx drive units. Frequency range 90-110 KHz. 2Hz discrimination. Crystal and Standardised centre frequency. Calibration accuracy ± 1% Ref. I.5. Price £135

**RECORDERS 4 PEN OSCILLOGRAPHS SOUTHERN INSTRUMENTS M942C.** 4 Channel fitted with 4 speed gear boxes giving 1, 5, 25, 100 m.m. per sec. Frequency response 0-55 Hz, sensitivity 0/m.m./M.A. Price £150  
**2 PEN OSCILLOGRAPH MR450** as per 4 Pen. Ref. I.2. Price £80 PLUS CARRIAGE

**E.M.I.** Portable L.F. Tape Recorder. Ex-service equipment consisting of Three Unit housed in transit cases (Tape Deck, Amplifier, P.S.U.). ¼ in. track speed 30 in., 15 in., 7½ in. and ¼ in. min. Price £75. Many control facilities. This is a good quality recorder.

**ELECTRONIC BROKERS LTD**



## EQUIPMENT AND COMPONENTS

### MEASURING INSTRUMENTS AND RECORDERS

#### MULTIMETER TYPE CT471B

Fully transistorized multi-range instrument for measurement of voltage up to 1000 MHz (1500 MHz with reduced accuracy) and current up to 2 kA and D.C. Resistance A.C. and D.C. voltage and current divided into 11 ranges.  
A.C./D.C. Volts 12mV-1200V.  
A.C./D.C. Current 12 micro A-1.2A.  
D.C. Resistance 5 ranges 0.1 ohm-1000 M ohm.  
R.F. Voltages 5 range 40mV to 4V.  
Battery powered. Offers in excellent condition. Tested before despatch. Complete with handbook. £54. Carriage 10/-.

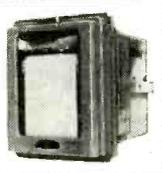


#### FACSIMILE RECORDERS

D649 K 18 in. Chart Recorder. Helix speed: 60, 90, 120 rev./min. Transmission speed: 1/2 in.; 15/16 in.; 1 1/4 in. per min. Scanning rate 96 lines/in.  
Ref. C.3..... Price £350. Completely overhauled + carriage

#### SINGLE PEN RECORDER

By Record Electrical (R3)  
3 in. chart, sensitivity 1 mA.  
Coil resistance 1.5k. Fully interchangeable gears available to make a wide range of chart speeds. 200/250V. Size: 8 1/2 x 6 1/2 in.  
Almost new—complete with chart and ink. List over £100.  
Our price ..... £49.50



#### POWER SUPPLY UNITS

O/P V	A	Sor	Input V	Make	Type	Dimensions W x H x L inches	Loca- tion	Ref. No.
6 adj								
4-16V	7	8	240	Contant	ELV700/6	6 1/2 x 5 1/2 x 11 1/2	876	36
6 adj								
3-16V	5	8	240	Contant	ELV600/6	4 1/2 x 7 x 12	876	37
4-15V	5	8	240	Advance	PM7	4 1/2 x 5 1/2 x 9 1/2	876	38
12	1	8	240	Farnell	88U12-1	4 x 6 x 10 1/2	876	39
± 150								
200 mA	8	240	Roband	B101/200	7 1/2 x 6 1/2 x 9	876	40	
28								
26-37	7	8	240	Contant	ES700/28	8 1/2 x 7 x 12	875	42
+30 300								
-20 mA	8	240	B.P.L.	3	19 x 8 1/2 x 12	875	41	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	879	66	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	878	70	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	878	57	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	878	56	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	878	59	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	878	58	
12	20	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	877	67	
6	16	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	879	54	
48	6	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 16	879	55	
30	7	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	879	62	
12	15	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	879	64	
12	12	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	874	60	
12	12	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	874	61	
20	6	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 13 1/2	874	65	
6	8	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 9 1/2	874	68	
6	8	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 9 1/2	874	63	
12	4	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 9 1/2	874	72	
12	4	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 9 1/2	874	71	
6	8	8	110	I.B.M.	Ex. comp. 6 x 5 1/2 x 9 1/2	874	69	
20	4 1/2	U/8		Electron-ics	SP110	8 x 6 x 13 1/2	877	43
-10 300 mA								
Do. Do. Do.	Do.	Do.	240	Do.	Do.		877	44
48	4	U/8	240	Advance	DC8	5 1/2 x 6 x 17	877	80
24	5	U/8	240	Advance	DC22	5 1/2 x 6 x 17	873	73
48	2	U/8	240	Advance	DC122	5 1/2 x 6 x 17	866	74
12/15	5	8	240	Advance	DCR12/12	5 1/2 x 8 x 17	873	53
6	20	8	240	Contant	R205	19 x 8 1/2 x 13 1/2	873	51
+ 6 10								
- 6	2	8	240	Contant	R206	19 x 7 x 12	873	47
28	20	8	240	Contant	R204	19 x 8 1/2 x 14	870	85
190-250								
350 mA	8	240	Airmec	705	19 x 12 x 8 1/2	872	82	

\* = Stab./Unstab.  
This is a small selection of our range. Further details on application.

### PRECISION POTENTIOMETERS

#### TEN TURN 360° ROTATION BRAND NEW (Ref. C5)

Res. Ohms	Linearity	Manufacturer	Model	Price
100/100/100		Beckman	A	£8.00
100	0.5	Beckman	A.S.	£3.00
200	0.5	Beckman	A	£3.00
500	0.1	Beckman	S	£3.50
500		Colvern	2501	£2.25
500		Foxes	PX4	£2.00
500		Colvern	2610	£2.50
500		Colvern	26/1000/11	£3.00
500	1.0	Relcon	HEL107-10	£2.25
1K		Relcon	HEL0710	£2.25
2K	0.5	Beckman	SA1101	£3.00
2K	0.25	Beckman	7218	£3.00
2K		Reliance	GPM15	£2.00
2K		General Controls	GPA15/4	£2.00
5K		Relcon	07-10	£2.50
5K		Colvern	CLR2503	£3.00
10K	0.5	Beckman	A	£3.00
10K	0.1	Beckman X	A	£3.50
10K		Beckman	SA950	£3.00
15K		Colvern	CLR26/1001	£3.50
18K		Beckman	CLR2402	£3.00
18K		Beckman	A	£3.00
25K	0.5	Helipot	SA337	£3.00
29K	0.05	Beckman	SA1244	£4.50
30K		Colvern	2402	£1.50
30K		Foxes	SA350	£3.00
30K	0.1	Beckman	A.88	£3.50
30K	0.5	Beckman	SA1692	£3.00
30K	0.25	Beckman	SA1679	£3.25
30K	1.0	Colvern	2402/1	£1.50
50K		Reliance	07-10	£2.25
50K		Colvern	07-5	£2.25
50K		Colvern	2503	£2.25
50K	X	Foxes	PX4	£2.25
50K	0.5	Beckman	A	£3.00
50K	0.1	Beckman	A	£3.50
100K/100K		Forl	A	£5.00
100K	0.1	Beckman	A	£3.50
100K	0.5	Beckman	A	£3.00
100K		Colvern	2501	£2.25
100K		Colvern	2610	£2.50
296K	0.1	Beckman	SA3902	£3.50
300K	0.1	Beckman	A	£3.50

#### THREE TURN 780° ROTATION

100/100	0.5	Beckman	C	£3.00
100/100		Beckman	Type C	£3.00
300		Beckman	9305	£2.25
1K		Foxes	PX2/H3	£2.25
10K	0.5	Beckman	C.S.	£2.25
20K/20K	0.1	Beckman	C.S.	£3.00
10K/10K	0.1	Beckman	C	£3.00
50K	0.5	Beckman	C.S.	£1.75

#### FIFTEEN TURN 5400° ROTATION

25K/25K		Beckman B	10 watts	£6.50
46K/46K		Beckman B	10 watts	£6.50

#### TWENTY TURN 7200° ROTATION

1 Meg		General Controls	PXM130	£4.00
50K		Reliance		£2.00

#### 156 TURN 56160° ROTATION

460		Kelvin Hughes	KTP0701	£9.50
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#### FIVE TURN 1800° ROTATION

200		Relcon	HEL07-05-P/11	£2.25
500		Colvern	CLR2605	£2.00
U1-5K		Colvern	CLR2605	£2.00

#### FIVE-AND-A-HALF TURN

500		Colvern	2405	£2.00
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#### TRIM POTENTIOMETERS (Ref. C7)

Just in Dobie McInnes Digital Analogue Converter and Plotter. (Ref. I.1)

Manufacturer	Value	Connection	Price
PAIGNTON	5 ohms	P.C.	50p
PAIGNTON	5 ohms	P.C.	50p
PAIGNTON	10 ohms	P.C.	50p
AMPHENOL	20 ohms		50p
AMPHENOL	50 ohms		50p
AMPHENOL	75 ohms	T.C. Turret Lugs	50p
AMPHENOL	100 ohms	P.C.	50p
MICROPOT	100 ohms	P.C.	50p
AMPHENOL	200 ohms	P.C.	50p
AMPHENOL	250 ohms	P.C.	50p
AMPHENOL	300 ohms	P.C.	50p
PAIGNTON	500 ohms	T.C.	50p
AMPHENOL	600 ohms	P.C.	50p
PAIGNTON	1 Kohms	P.C.	75p
AMPHENOL	2 Kohms	P.C.	75p
PAIGNTON	2.5 Kohms	P.C.	75p
AMPHENOL	2.5 Kohms	T.C.	75p
AMPHENOL	3 Kohms	T.C.	75p
BOURNES	5 Kohms	Stud Connection	75p
BOURNES	5 Kohms	Flying lead	75p
AMPHENOL	10 Kohms	P.C.	£1
AMPHENOL	25 Kohms	P.C.	£1
AMPHENOL	30 Kohms	T.C.	£1

#### MONOCHROMATIC LIGHT "LAPMASTER"

110/230V 1-ph. 50 Hz. Light area: 1 1/2 in. x 9/16 in. £15.00. Carriage £2.00.

#### SOUND ANALYSER

General Radio Co. Type 760-A. Portable. Battery powered. Designed for use with Type 759 sound level meters, but can be used with any other microphone or vibration pick-up and amplifier with suitable characteristics. Supplied less microphone. £50. Carriage extra.

#### MARCONI TF899

20mV-2V A.C. 3 ranges. 50Hz-100MHz. Detected O/p for modulation monitoring RF probe. Mains P.S.U. Overhauled. £25. P. & P. £0.50.

#### PHILLIPS GM6020

100 micro V-10V, input 1. 10mV-1000V input 2. 100 pA-10 micro A. Accuracy ± 3%. Input 2 1 M ohm, input 1, 100 M ohm, input 2. Recorder output. Working order £65. P. & P. £1.

#### VHF ADMITTANCE BRIDGE

Wayne Kerr B801A. 1-100 MHz. Conductance 0-100 millimhos. Capacitance 0-230 pF and 0 to ±230 pF. £120 (40% of new price). Also B901. Indicates parallel components of conductance and positive or negative capacitance for lines, antennas and feeders. 0-100mMho. 0. to ± 75 pF and -75 pF. Accuracy 2% up to 250 MHz. £115 (40% of new price).

#### FENLOW LOW FREQUENCY ANALYSER

0.3 Hz to 1 K Hz. Power density 0-10. Bandwidth switching range. .06: 0.3: 1.5: 7.5: 37.5 Hz. Price £275.

#### TWENTY MILLION MEGOHMMETER

E.I. Model 29A. Test voltage 85 and 500V. S/C Current less than 4 mA 30M ohm-20 x 10^4 ohm. Charging Delay 1 1/2 secs. Mains input. £75. Carriage £2.50.

#### NEW ELECTRO PNEUMATIC TRANSDUCER TRANSMITTER

Taylor. Cat. No. XX701 TF13. Input -50-0-50 Ma. Output 3-15 PSI. Spec. 670. Coil 3 ohms. This precision transducer accurately controls air pressure by a varying electrical signal. £50. P. & P. included.

#### R.C. OSCILLATOR

Solartron Type CO 1004-2. 10Hz-1 MHz in 5 ranges. O/P level adjustment. £40. P. & P. £1. Also available Type CO 1004. £30. P. & P. £1.

#### PORTABLE FREQUENCY METERS

TF1026/1. A direct reading absorption meter, employing a concentric line closed at one end and turned by variable capacitor at the other end of the line, giving a frequency range: 250 MHz-300 MHz, on an almost linear scale approx. 9 in. in length. Complete in polished wooden case. Price £17.50. Carriage extra.

#### DIGITAL INDICATORS KGM Type M3

A neat compact indicator providing selective display 0-9. Fig. height 18 mm. panel mounting, 6 mm. tubular midjet flange lamps. Supplied with 28 v. bulbs. Finished matt black anodized. W. 1 in., H. 2 in. Wt. 4 ozs. Price £3.25. P. & p. Free.

#### MODEL 1706 VISICORDER

In almost new condition. This direct reading U/V Recorder can record up to 6 channels simultaneously from D.C. 5000 Hz at writing speed of 30000 m ohs/sec. Recording range: D.C.-5000 Hz. Paper width: 4 1/2 in. wide. Optical Arm: 19 cm. Paper Speeds: Eight speeds from 0.25-32 in./sec. and 6-300 mm/sec. Dimensions: H. 10 1/2 in., W. 12 in., Depth 14 in. Complete with 4 3k Hz Galvos. £400

#### BRAND NEW CAPACITOR REVERSIBLE SINGLE PHASE PARVALUX MOTORS

230/250 v. 50 Hz 2,800 r.p.m. 1/30 h.p. Cont. rated. 3/8 in. shaft dia. x 3 1/2 in. long. Foot mounting. Weight 6 lb. £5.75 post free.

#### COAXIAL LINE OSCILLATOR

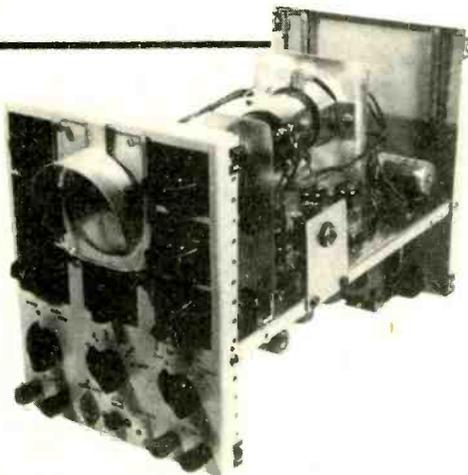
By Saunders. Type CLC 7-12. The Oscillator is adjustable from 7-12 MHz. A high reset accuracy with no

# EXTRA

in April issue of "Practical Wireless" OUT NOW

## 8-PAGE GUIDE TO TEST INSTRUMENTS

This illustrated eight-page supplement is a comprehensive guide to test gear now on the market for use in testing radio and audio equipment. Though complete in itself, it will also provide a useful introduction to an important new series on servicing by H. W. Hellyer and Gordon J. King, which starts in the following month's issue.



### WORKSHOP OSCILLOSCOPE

This *Practical Wireless* design enables you to see what is going on inside a circuit, including the examination of waveforms, distortion and pulses. You can make this versatile test instrument — readily available c.r.t. and other components. Even the case is a stock item. Read all about it in the April issue, out now.



### HIGH IMPEDANCE VOLTMETER

Winner of the *Practical Wireless* Project Autumn Trophy with ranges of 1-10-100-1000V and input impedance of at least 3M $\Omega$ . The cost is small, a 50 $\mu$ A meter and three 2N2926 transistors forming the basis of operation of this ingenious project.

Full constructional details in the April *Practical Wireless*

*Also other interesting constructional features*

# PRACTICAL WIRELESS

April issue out now—17½p

# ELECTROVALUE

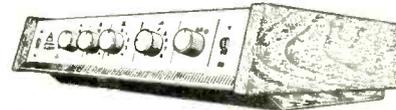
EVERYTHING BRAND NEW TO SPEC • LARGE STOCKS • NO SURPLUS

## BARGAINS IN NEW SEMI-CONDUCTORS

MANY AT NEW REDUCED PRICES • ALL POWER TYPES WITH FREE INSULATING SETS

40361	55p	2N2905	44p	2N4291	15p	BC148	14p	BFX87	29p
40362	68p	2N2905A	47p	2N4292	15p	BC149	15p	BFX88	26p
2N696	20p	2N2924	20p	AC107	46p	BC153	19p	BFY50	23p
2N697	22p	2N2925	22p	AC126	20p	BC154	28p	BFY51	20p
2N706	12p	2N2926	11p	AC127	20p	BC157	19p	BFY52	23p
2N930	29p	2N3053	27p	AC128	20p	BC158	17p	B5X20	16p
2N1131	36p	2N3055	75p	AC153K	25p	BC159	18p	C407	17p
2N1132	40p	2N3702	13p	AC176	27p	BC167	13p	MCI140	25p
2N1302	19p	2N3703	13p	ACY20	20p	BC168	11p	MPS6531	35p
2N1303	19p	2N3704	13p	ACY22	16p	BC169	13p	MPS6534	30p
2N1304	23p	2N3705	13p	AD140	56p	BC177	17p	NKT211	25p
2N1305	23p	2N3706	13p	AD142	50p	BC178	15p	NKT212	25p
2N1306	33p	2N3707	13p	AD149	60p	BC179	17p	NKT214	23p
2N1307	33p	2N3708	13p	AD161	40p	BC182L	13p	NKT274	18p
2N1308	36p	2N3709	13p	AD162	40p	BC183L	11p	NKT403	65p
2N1309	36p	2N3710	13p	AF114	30p	BC184L	13p	NKT405	79p
2N1613	23p	2N3711	13p	AF115	30p	BC212L	25p	OC71	29p
2N1711	26p	2N3819	35p	AF117	28p	BC213L	25p	OC81	25p
2N1893	54p	2N3904	35p	AF124	30p	BC214L	25p	OC81D	25p
2N2147	95p	2N3906	35p	AF127	28p	BCY70	19p	ZTX300	17p
2N2218	34p	2N4058	20p	AF139	48p	BCY71	33p	ZTX301	17p
2N2218A	43p	2N4059	20p	AF239	49p	BCY72	15p	ZTX302	22p
2N2219	38p	2N4060	20p	ASY26	27p	BF115	23p	ZTX303	22p
2N2219A	53p	2N4061	20p	ASY28	27p	BF167	27p	ZTX304	33p
2N2270	62p	2N4062	20p	BC107	14p	BF173	31p	ZTX500	25p
2N2369A	19p	2N4124	18p	BC108	12p	BF194	17p	ZTX501	25p
2N2483	35p	2N4126	27p	BC109	14p	BF195	18p	ZTX502	30p
2N2484	42p	2N4284	15p	BC125	15p	BFX29	31p	ZTX503	25p
2N2646	54p	2N4286	15p	BC126	22p	BFX84	25p	ZTX504	60p
2N2904A	42p	2N4289	15p	BC147	15p	BFX85	34p		

## PEAK SOUND PRODUCTS



ENGLEFIELD  
12+12  
AMPLIFIER

Stereo amplifier in modular kit form 12 watts RMS per channel into 15Ω. £38.45. Cabinet kit only £6. These prices nett. As reviewed in *Hi Fi Sound* and other important journals.



BAXANDALL  
SPEAKER SYSTEM

Designed by Peter Baxandall. Superb reproduction for its size. Handles 10 watts with ease. Uses ELAC 15Ω 59RM109 speaker unit. Kit £13.90 nett; built £19.40 nett.

## MAINLINE AMPLIFIER KITS

RC A/SGS designed main amplifier kits. Input sensitivity 500-700mV for full output into 8Ω.

Power	Kit price including components	Suitable unreg. power supply kit
12W	£8.40 nett	£4.82
25W	£9.75 nett	£5.92
40W	£10.50 nett	£6.03
70W	£12.60 nett	£6.87

## 30 WATT BAILEY AMPLIFIER PARTS

Sensitivity 1.2V for full output into 8Ω. Transistors and PCB for one channel £6.46. Transistors and PCBs for two channels £12.92. Capacitors and resistors (metal oxide), £2.00 per channel. Complete unregulated power supply pack, £4.75. Suitable heat sink 10DN space 400c, 55p.

## INTEGRATED CIRCUITS

PLESSEY SL403A 3 wats into 7.5 ohms. Application data, 10p. Price per unit, nett £2.10

SINCLAIR IC.10 as advertised, complete with instructions and applications manual £2.95 nett. Components pack for stereo inc. transformer, controls, etc., £4.75 nett.

S-DeCs PUT AN END TO BIRDS NESTING. Components just plug in—saves time—allows re-use of components. S-Dec (70 points), £1.00. Complete T-Dec, may be temperature-cycled (208 points), £2.50. Also μ-Decs and IC carriers.

MEDIUM RANGE ELECTROLYTICS. Axial leads: 50/50, 9p; 100/25, 9p; 100/50, 13p; 250/25, 13p; 250/50, 19p; 500/25, 19p; 500/50, 21p; 1000/25, 20p; 1000/50, 30p; 2000/25, 30p; 2000/50 48p.

SMALL ELECTROLYTICS. Axial leads: 4.7/10, 4.7/25, 5/50, 5p each; 10/10, 10/25, 10/50, 33/10, 50/10, 5p each; 25/25, 25/50, 47/25, 100/10, 220/10, 6p each.

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## RESISTORS

Code	Power	Tolerance	Range	Values available	1 to 9	10 to 99	100 up
C	1/20W	5%	82Ω-220KΩ	E12	7	6.5	6
C	1/8W	5%	4.7Ω-330KΩ	E24	1.5	0.8	0.7
C	1/4W	10%	4.7Ω-10MΩ	E12	1.5	0.8	0.7
C	1/2W	5%	4.7Ω-10MΩ	E24	1.2	1	0.9
C	1W	10%	4.7Ω-10MΩ	E12	2.5	2	1.9
MO	1/2W	2%	10Ω-1MΩ	E24	4	3.5	3
WW	1W	10%	0.22Ω-3.9Ω	E12	7	7	6
WW	3W	5%	12Ω-10KΩ	E12	7	7	6
WW	7W	5%	12Ω-10KΩ	E12	9	9	8

Codes: C = carbon film, high stability, low noise. MO = metal oxide, ElectroSil TR5, ultra low noise. WW = wire wound, Plessey.

Values: E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

ZENER DIODES 5% full range E24 values: 400mV: 2.7V to 30V, 15p each; 1W: 6.8V, to 82V, 27p each; 1.5W: 4.7V to 75V, 60p each. Clip to increase 1.5W rating to 3 watts (type 266F), 4p.

CARBON TRACK POTENTIOMETERS, long spindles. Double wiper ensures minimum noise level. Single gang linear 220Ω to 2.2MΩ, 12p; Single gang log, 4.7kΩ to 2.2MΩ, 12p; Dual gang linear, 4.7kΩ to 2.2MΩ, 42p; Dual gang log, 4.7kΩ to 2.2MΩ, 42p; Log/antilog, 10k, 47k, 1MΩ only 42p; Dual antilog, 10k only, 42p. Any type with 1/2A D.P. mains switch, extra 12p. Please note: only decades of 10, 22 and 47 are available within ranges quoted.

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Tuner 100mV into 100K ohms.

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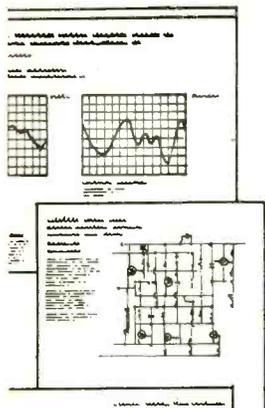
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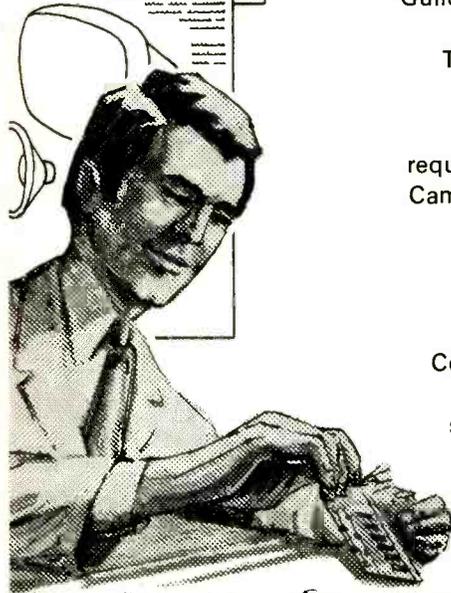


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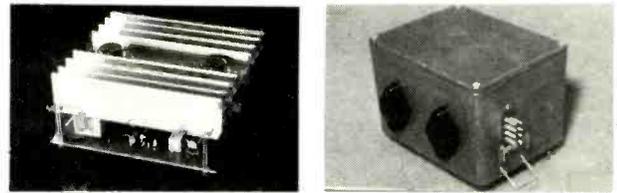


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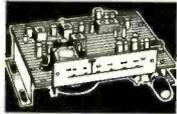
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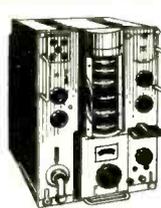


Oscillator Test No. 2. A high quality precision instrument made for the Ministry by Airmeq. Frequency coverage 20/80 Mc/s. AM/CW/FM. Incorporates precision dial, level meter, precision attenuator 1µV-100µV. Operation from 12 volt D.C. or 0/110/200/250 v. A.C. Size 12 x 8 1/2 x 9in. Supplied in brand new condition complete with all connectors, fully tested, £45. Carr. £1.

**AVO CT.38 ELECTRONIC MULTIMETERS**

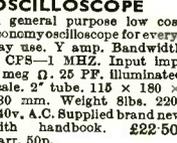
High quality 97 range instrument which measures A.C. and D.C. Voltage, Current, Resistance and Power Output Ranges D.C. volts 250 mV-10,000V. (10 meg Ω-110 meg Ω input). D.C. current 10µA-25 amps. Ohms. 0-1,000 meg Ω. A.C. volt 100mV-250V (with R.F. measuring head up to 250 Mc/s). A.C. current 10µA-25 amps. Power output 50 micro-watts-5 watts. Operation 0/110/200/250V. A.C. Supplied in perfect condition complete with circuit lead and R.F. probe. £25. Carr. 75p.

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High quality 10 valve receiver manufactured by Murphy. Coverage in 5 bands 150-300 Kc/s, 560 Kc/s-1.5Mc/s, 3-9-30.5 Mc/s, I.F. 500 Kc/s. Incorporates 2 R.F. and 3 I.F. stages, bandpass filter, noise limiter, crystal controlled B.F.O. calibrator I.F. output, etc. Built-in speaker, output for phones. Operation 150/230 volt A.C. Size 19 1/2 x 13 1/2 x 16in. Weight 114lb. Offered in good working condition. £22.50. Carr. £1.50. With circuit diagrams. Also available B41 L.F. version of above. 15 Kc/s-700 Kc/s. £17.50. Carr. £1.50.

**TO-2 PORTABLE OSCILLOSCOPE**



A general purpose low cost economy oscilloscope for everyday use. Y amp. Bandwidth 2 CPS-1 MHz. Input Imp. 2 meg Ω. 25 PF. Illuminated scale. 2" tube, 115 x 180 x 230 mm. Weight 8lbs. 220/240V. A.C. Supplied brand new with handbook. £22.50. Carr. 50p.

**TO-3 PORTABLE OSCILLOSCOPE. 3" TUBE**



Y amp. Sensitivity. 1v p-p/CM. Bandwidth 1.5 cps 1-5 MHz. Input Imp. 2 meg Ω. 25 PF. X amp sensitivity. .9v p-p/CM. bandwidth 1.5 cps-800 KHz. Input Imp. 2 meg Ω. 20 PF. Time base. 5 ranges 19 cps-300 KHz. Synchronisation. Internals/external. Illuminated scale. 140 x 215 x 330 mm. Weight 15 1/2 lbs. 220/240 V. A.C. Supplied brand new with handbook £27.50. Carr. 50p.

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100µA	£3.37	300V. D.C.	£2.97
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500µA	£3.25	5 amp. D.C.	£2.97
1mA	£3.12	300V. A.C.	£2.97
	£2.97	VU Meter	£3.75

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**TYPE S-80**  
80 mm. square fronts

50µA	£3.12	50V. D.C.	£2.47
50-0-50µA	£2.97	300V. D.C.	£2.47
100µA	£2.97	1 amp. D.C.	£2.47
100-0-100µA	£2.97	5 amp. D.C.	£2.47
500µA	£2.82	300V. A.C.	£2.82
1mA	£2.47	VU Meter	£3.37
20V. D.C.	£2.47		

**"SEW" CLEAR PLASTIC METERS**

**Type MR.52P. 4 1/2in. x 4 1/2in. fronts.**

10mA	£2.80	200mA	£2.80
50mA	£2.80	500mA	£2.80
100µA	£2.80	1 amp.	£2.80
500µA	£2.80	5 amp.	£2.80
1mA	£2.80	15 amp.	£2.80
5mA	£2.80	30 amp.	£2.80
10mA	£2.80	20V. D.C.	£2.80
50mA	£2.80	50V. D.C.	£2.80
100-0-100µA	£2.80	150V. D.C.	£2.80
500µA	£2.80	300V. D.C.	£2.80
1mA	£2.80	15V. A.C.	£2.80
5mA	£2.80	300V. A.C.	£2.80
10mA	£2.80	5 Meter 1mA	£2.97
50mA	£2.80	VU Meter	£3.60
100mA	£2.80	1 amp. A.C.*	£2.80
500mA	£2.80	5 amp. A.C.*	£2.80
1mA	£2.80	10 amp. A.C.*	£2.80
5mA	£2.80	20 amp. A.C.*	£2.80
	£2.80	30 amp. A.C.*	£2.80

**Type MR.38P. 1 21/32in. square fronts.**

200mA	£1.37	500mA	£1.37
500mA	£1.37	1 amp.	£1.37
1 amp.	£1.37	2 amp.	£1.37
2 amp.	£1.37	5 amp.	£1.37
5 amp.	£1.37	10 amp.	£1.37
10 amp.	£1.37	3V. D.C.	£1.37
3V. D.C.	£1.37	10V. D.C.	£1.37
10V. D.C.	£1.37	15V. D.C.	£1.37
15V. D.C.	£1.37	20V. D.C.	£1.37
20V. D.C.	£1.37	100V. D.C.	£1.37
100V. D.C.	£1.37	150V. D.C.	£1.37
150V. D.C.	£1.37	300V. D.C.	£1.37
300V. D.C.	£1.37	500V. D.C.	£1.37
500V. D.C.	£1.37	750V. D.C.	£1.37
750V. D.C.	£1.37	1 amp. A.C.*	£1.37
1 amp. A.C.*	£1.37	5 amp. A.C.*	£1.37
5 amp. A.C.*	£1.37	10 amp. A.C.*	£1.37
10 amp. A.C.*	£1.37	20 amp. A.C.*	£1.37
20 amp. A.C.*	£1.37	30 amp. A.C.*	£1.37
30 amp. A.C.*	£1.37	VU Meter	£2.10

**Type MR.52P. 2 1/2in. square fronts.**

50µA	£3.10	10V. D.C.	£2.00
50-0-50µA	£2.80	20V. D.C.	£2.00
100µA	£2.80	50V. D.C.	£2.00
100-0-100µA	£2.37	300V. D.C.	£2.00
500µA	£2.25	15V. A.C.	£2.00
1mA	£2.00	300V. A.C.	£2.00
5mA	£2.00	8 Meter 1mA	£2.10
10mA	£2.00	VU Meter	£3.10
50mA	£2.00	1 amp. A.C.*	£2.00
100mA	£2.00	5 amp. A.C.*	£2.00
500mA	£2.00	10 amp. A.C.*	£2.00
1 amp.	£2.00	20 amp. A.C.*	£2.00
5 amp.	£2.00	30 amp. A.C.*	£2.00

**Type MR.45P. 2in. square fronts.**

50µA	£2.25	5 amp.	£1.50
50-0-50µA	£2.10	10V. D.C.	£1.50
100µA	£2.10	20V. D.C.	£1.50
100-0-100µA	£1.87	50V. D.C.	£1.50
200µA	£1.87	300V. D.C.	£1.50
500µA	£1.80	15V. A.C.	£1.50
500-0-500µA	£1.50	300V. A.C.	£1.50
1mA	£1.50	500V. D.C.	£1.50
5mA	£1.50	8 Meter 1mA	£1.87
10mA	£1.50	VU Meter	£2.25
50mA	£1.50	1 amp. A.C.*	£1.50
100mA	£1.50	5 amp. A.C.*	£1.50
500mA	£1.50	10 amp. A.C.*	£1.50
1 amp.	£1.50	20 amp. A.C.*	£1.50
	£1.50	30 amp. A.C.*	£1.50

**"SEW" BAKELITE PANEL METERS**

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500mA	£1.75	500µA	£1.75
1 amp.	£1.75	100µA	£1.75
5 amp.	£1.75	50µA	£1.75
15 amp.	£1.75	200µA	£1.75
30 amp.	£1.75	100-0-100µA	£1.75
50 amp.	£1.75	500µA	£1.75
5V. D.C.	£1.75	500-0-500µA	£1.75
10V. D.C.	£1.75	1mA	£1.75
20V. D.C.	£1.75	5mA	£1.75
50V. D.C.	£1.75	10mA	£1.75
100V. D.C.	£1.75	50mA	£1.75
15V. A.C.	£1.75	100mA	£1.75
300V. A.C.	£1.75	500mA	£1.75
5 amp. A.C.*	£1.75	1 amp.	£1.75
10 amp. A.C.*	£1.75		
20 amp. A.C.*	£1.75		
30 amp. A.C.*	£1.75		
VU Meter	£3.10		

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**Type ED 107**  
Size overall 100mm x 90mm x 108mm

A new range of high quality moving coil instruments ideal for school experiments and other bench applications. The meter movement is easily accessible to demonstrate internal working.

50µA	£4.50	20V. D.C.	£3.97
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1mA	£3.97	300V. D.C.	£3.97
50-0-50µA	£4.25	Dual range	
500µA	£3.97	500mA/5A d.c.	£4.25
1mA	£3.97	5V/50V A.C.	£4.25
5A d.c.	£3.97	Triple range	
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Can be panel or bench mounted. Basic meter measures 1 volt D.C. but can be used to measure a wide range of AC and DC volt, current and ohms with optional plug in cards.

Specification:  
Accuracy: ± 0.2, ± 1 digit.  
Resolution: 1mV.  
Number of digits: 3 plus fourth overrange digit.  
Overrange: 100% (up to 1-999)  
Input impedance: 1000 Meg ohm.  
Measuring cycle: 1 per second.  
Adjustment: Automatic zeroing, full scale adjustment against an internal reference voltage.  
Overload: to 100V. D.C.  
Input: Fully floating (3 poles).  
Input power: 110-230V. A.C. 50/60 cycles.  
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2N929	22p	BC113	25p	NKT405	75p	6AU6	25p
2N930	22p	BC113	25p	NKT405	75p	6B4	30p
2N1131	30p	BC113	25p	NKT405	75p	6BA6	25p
2N1132	30p	BC116	40p	OA5	20p	6BE6	30p
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2N2218	30p	BCY33	25p	OC23	60p	6F1	60p
2N2219	32p	BCY34	30p	OC24	60p	6F1	60p
2N2388	17p	BCY38	40p	OC25	37p	6F1	60p
2N2389	20p	BCY42	15p	OC26	25p	6F1	60p
2N2394	25p	BCY43	20p	OC28	62p	6F1	60p
2N2484	35p	BCY71	30p	OC29	62p	6F1	60p
2N2613	35p	BCY72	15p	OC35	50p	6F1	60p
2N2646	50p	BCZ11	40p	OC36	62p	6F1	60p
2N2904	30p	BD121	65p	OC41	25p	6F1	60p
2N2923	17p	BD123	80p	OC42	30p	6F1	60p
2N2924	17p	BD124	80p	OC43	17p	6F1	60p
2N2925	17p	BD125	25p	OC45	15p	6F1	60p
2N2926	12p	BD126	25p	OC46	27p	6F1	60p
2N2927	12p	BD127	25p	OC70	12p	6F1	60p
2N2928	12p	BD128	25p	OC71	15p	6F1	60p
2N2929	12p	BD129	25p	OC72	25p	6F1	60p
2N3054	25p	BD130	25p	OC73	30p	6F1	60p
2N3055	25p	BD131	25p	OC74	30p	6F1	60p
2N3056	25p	BD132	25p	OC75	25p	6F1	60p
2N3057	25p	BD133	25p	OC76	25p	6F1	60p
2N3058	25p	BD134	25p	OC77	40p	6F1	60p
2N3059	25p	BD135	25p	OC78	20p	6F1	60p
2N3060	25p	BD136	25p	OC81	25p	6F1	60p
2N3061	25p	BD137	25p	OC81D	20p	6F1	60p
2N3062	25p	BD138	25p	OC83	25p	6F1	60p
2N3063	25p	BD139	25p	OC84	25p	6F1	60p
2N3064	25p	BD140	25p	OC139	25p	6F1	60p
2N3065	25p	BD141	25p	OC140	37p	6F1	60p
2N3066	25p	BD142	25p	OC169	20p	6F1	60p
2N3067	25p	BD143	25p	OC170	25p	6F1	60p
2N3068	25p	BD144	25p	OC171	30p	6F1	60p
2N3069	25p	BD145	25p	OC200	40p	6F1	60p
2N3070	25p	BD146	25p	OC201	60p	6F1	60p
2N3071	25p	BD147	25p	OC202	75p	6F1	60p
2N3072	25p	BD148	25p	OC203	40p	6F1	60p
2N3073	25p	BD149	25p	OC204	40p	6F1	60p
2N3074	25p	BD150	25p	OC205	62p	6F1	60p
2N3075	25p	BD151	25p	OC207	75p	6F1	60p
2N3076	25p	BD152	25p	OC207	75p	6F1	60p
2N3077	25p	BD153	25p	OC211	97p	6F1	60p
2N3078	25p	BD154	25p	OC212	50p	6F1	60p
2N3079	25p	BD155	25p	OC213	50p	6F1	60p
2N3080	25p	BD156	25p	OC214	50p	6F1	60p
2N3081	25p	BD157	25p	OC215	50p	6F1	60p
2N3082	25p	BD158	25p	OC216	50p	6F1	60p
2N3083	25p	BD159	25p	OC217	50p	6F1	60p
2N3084	25p	BD160	25p	OC218	50p	6F1	60p
2N3085	25p	BD161	25p	OC219	50p	6F1	60p
2N3086	25p	BD162	25p	OC220	50p	6F1	60p
2N3087	25p	BD163	25p	OC221	50p	6F1	60p
2N3088	25p	BD164	25p	OC222	50p	6F1	60p
2N3089	25p	BD165	25p	OC223	50p	6F1	60p
2N3090	25p	BD166	25p	OC224	50p	6F1	60p
2N3091	25p	BD167	25p	OC225	50p	6F1	60p
2N3092	25p	BD168	25p	OC226	50p	6F1	60p
2N3093	25p	BD169	25p	OC227	50p	6F1	60p
2N3094	25p	BD170	25p	OC228	50p	6F1	60p
2N3095	25p	BD171	25p	OC229	50p	6F1	60p
2N3096	25p	BD172	25p	OC230	50p	6F1	60p
2N3097	25p	BD173	25p	OC231	50p	6F1	60p
2N3098	25p	BD174	25p	OC232	50p	6F1	60p
2N3099	25p	BD175	25p	OC233	50p	6F1	60p
2N3100	25p	BD176	25p	OC234	50p	6F1	60p
2N3101	25p	BD177	25p	OC235	50p	6F1	60p
2N3102	25p	BD178	25p	OC236	50p	6F1	60p
2N3103	25p	BD179	25p	OC237	50p	6F1	60p
2N3104	25p	BD180	25p	OC238	50p	6F1	60p
2N3105	25p	BD181	25p	OC239	50p	6F1	60p
2N3106	25p	BD182	25p	OC240	50p	6F1	60p
2N3107	25p	BD183	25p	OC241	50p	6F1	60p
2N3108	25p	BD184	25p	OC242	50p	6F1	60p
2N3109	25p	BD185	25p	OC243	50p	6F1	60p
2N3110	25p	BD186	25p	OC244	50p	6F1	60p
2N3111	25p	BD187	25p	OC245	50p	6F1	60p
2N3112	25p	BD188	25p	OC246	50p	6F1	60p
2N3113	25p	BD189	25p	OC247	50p	6F1	60p
2N3114	25p	BD190	25p	OC248	50p	6F1	60p
2N3115	25p	BD191	25p	OC249	50p	6F1	60p
2N3116	25p	BD192	25p	OC250	50p	6F1	60p
2N3117	25p	BD193	25p	OC251	50p	6F1	60p
2N3118	25p	BD194	25p	OC252	50p	6F1	60p
2N3119	25p	BD195	25p	OC253	50p	6F1	60p
2N3120	25p	BD196	25p	OC254	50p	6F1	60p
2N3121	25p	BD197	25p	OC255	50p	6F1	60p
2N3122	25p	BD198	25p	OC256	50p	6F1	60p
2N3123	25p	BD199	25p	OC257	50p	6F1	60p
2N3124	25p	BD200	25p	OC258	50p	6F1	60p
2N3125	25p	BD201	25p	OC259	50p	6F1	60p
2N3126	25p	BD202	25p	OC260	50p	6F1	60p
2N3127	25p	BD203	25p	OC261	50p	6F1	60p
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2N3133	25p	BD209	25p	OC267	50p	6F1	60p
2N3134	25p	BD210	25p	OC268	50p	6F1	60p
2N3135	25p	BD211	25p	OC269	50p	6F1	60p
2N3136	25p	BD212	25p	OC270	50p	6F1	60p
2N3137	25p	BD213	25p	OC271	50p	6F1	60p
2N3138	25p	BD214	25p	OC272	50p	6F1	60p
2N3139	25p	BD215	25p	OC273	50p	6F1	60p
2N3140	25p	BD216	25p	OC274	50p	6F1	60p
2N3141	25p	BD217	25p	OC275	50p	6F1	60p
2N3142	25p	BD218	25p	OC276	50p	6F1	60p
2N3143	25p	BD219	25p	OC277	50p	6F1	60p
2N3144	25p	BD220	25p	OC278	50p	6F1	60p
2N3145	25p	BD221	25p	OC279	50p	6F1	60p
2N3146	25p	BD222	25p	OC280	50p	6F1	60p
2N3147	25p	BD223	25p	OC281	50p	6F1	60p
2N3148	25p	BD224	25p	OC282	50p	6F1	60p
2N3149	25p	BD225	25p	OC283	50p	6F1	60p
2N3150	25p	BD226	25p	OC284	50p	6F1	60p
2N3151	25p	BD227	25p	OC285	50p	6F1	60p
2N3152	25p	BD228	25p	OC286	50p	6F1	60p
2N3153	25p	BD229	25p	OC287	50p	6F1	60p
2N3154	25p	BD230	25p	OC288	50p	6F1	60p
2N3155	25p	BD231	25p	OC289	50p	6F1	60p
2N3156	25p</						

**RUSSIAN CI-18 DOUBLE BEAM OSCILLOSCOPE**  
5 Mc/s Pass Band. Separate Y1 and Y2 amplifiers. Rectangular 5 in. x 4 in. C.R.T. Calibrated triggered sweep from 2  $\mu$ /sec. to 100 milli-sec. per cm. Free running time base 50 c/s-1 mc/s. Built-in time base calibrator and amplitude calibrator. Supplied complete with all accessories and instruction manual. £87 Carr. paid.



**MARCONI CT44 TF956 AF ABSORPTION WATTMETER**  
1  $\mu$ /watt to 6 watts. £20. Carr. £1.



**TE111 DECADE RESISTANCE ATTENUATOR**  
Variable range 0-111 db. Connections, Unbalanced T and Bridge T. Impedance 600 ohms. Range (0.1 db x 10) + (1 db x 10) + 10 + 20 + 30 + 40 db. Frequency: DC to 200 KHZ (-3db). Accuracy: 0.05 db. + indication db x 0.01. Maximum input less than 4 watts (50 volts). Built in 600  $\Omega$  load resistance with internal/external switch. Brand new £27.50 P. & P. 25p.



**BELCO AF-5A SOLID STATE SINE SQUARE WAVE C.R. OSCILLATOR**  
Sine 18-200,000 Hz; Square 18-500,000 Hz. Output max. +10 dB (10 K ohms). Operation internal batteries. Attractive 2-tone case 7 1/2 in. x 5 in. x 2 1/2 in. Price £17.50 Carr. 17 1/2p.



**TE-16A TRANSISTORISED SIGNAL GENERATOR**  
5 Ranges 400 kHz-30 MHz. An inexpensive instrument for the handyman. Operates on 9 v. battery. Wide easy to read scale. 800 kHz modulation 5 1/2 in. x 5 1/2 in. x 3 1/2 in. Complete with instructions and leads. £7.97 P. & P. 20p.



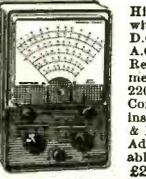
**BELCO DA-20 SOLID STATE DECADE AUDIO OSCILLATOR**  
New high-quality portable instrument. Sine 1 Hz to 100 KHZ. Square 20 Hz to 20 KHZ. Output max. +10 db (10 K ohms). Operation 220/240 v. A.C. Size 215 mm x 150 mm x 120 mm. Price £27.50 Carr. 25p.



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**TE-65 VALVE VOLTMETER**  
High quality instrument with 28 ranges. D.C. volts 1.5-1,500 v. A.C. volts 1.5-1,500 v. Resistance up to 1,000 megohms. 220/240v. A.C. operation. Complete with probe and instructions £17.50. P. & P. 30p. Additional Probes available: R.F. £1.75 H.V. £2.12.

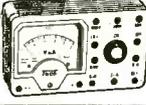


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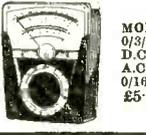
**MODEL TE-200 20,000 O.P.V. Mirror scale, overload protection 0/5/25/125/1,000 V.D.C. 0/10/50/250/1,000 V.A.C. 0/50  $\mu$ A/250 mA. 0/60K/6 meg. + 20 to + 62 db. £3.75 P. & P. 15p**



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**TMK MODEL TW-50K 46 ranges, mirror scale. 50K  $\Omega$ /Volt D.C. 5K  $\Omega$ /Volt A.C. D.C. Volts: .125, .25, 1.25, 2.5, 5, 10, 25, 50, 125, 250, 500, 1000V. A.C. Volts: 1.5, 3, 5, 10, 25, 50, 125, 250, 500, 1000V. D.C. Current: 25, 50  $\mu$ A, 2.5, 5, 25, 50, 250, 500mA. 1 MEG. 10 MEG. Decibels: -20 to +81.5 db. £8.87 P. & P. 17 1/2p**



**TE-900 20,000  $\Omega$ /VOLT GIANT MULTIMETER. Mirror scale and overload protection. 6 in. full view meter. 2 colour scale. 0/2.5/10/25/100/500/1,000 v. A.C. 0/25/125/10/50/250/1,000/5,000 v. D.C. 0/50  $\mu$ A/0/10/100/500mA/10 amp. D.C. 0/2K/200K/20 MEG. OHM. £15 P. & P. 28p**



**MODEL 5025 57 Ranges, Giant 5 1/2 in. Meter, Polarity Reverse Switch. Sensitivity: 50K  $\Omega$ /Volt D.C. 5K  $\Omega$ /Volt A.C. D.C. Volts: .125, .25, 1.25, 5, 10, 25, 50, 125, 250, 500, 1,000V. A.C. Volts: 1.5, 3, 5, 10, 25, 50, 125, 250, 500, 1,000V. D.C. Current: 25, 50  $\mu$ A, 2.5, 5, 25, 50, 250, 500mA. 10 MEG. Decibels: -20 to +85 dB £12.50 P. & P. 17 1/2p.**



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Full capabilities for measuring A, B and 100. NPN or PNP. Equally adaptable for checking diodes. Supplied complete with instructions, battery and leads. £6.97. P. & P. 15p.



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**MODEL TE-300 30,000 O.P.V. Mirror scale, overload protection 0/6/3/15/60/300/1,200 V.D.C. 0/6/30/120/600/1,200 V.A.C. 0/30  $\mu$ A/30mA/60mA/300mA/600mA. 0/8K/80K/800K/8 meg. -20 to + 63 db. £5.97. P. & P. 15p.**



**MODEL TE-12. 20,000 O.P.V. 0/0.6/6/30/120/600/1,200/3,000/6,000 v. D.C. 0/6/30/120/600/1,200 v. A.C. 0/60  $\mu$ A/6/60/600 mA. 0/6K/60K/6MEG. 60 Meg.  $\Omega$ . 50 PF. 2 MFD. £5.97. P. & P. 17 1/2p.**



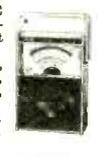
**MODEL 600 30,000 O.P.V. with overload protection, mirror scale. 0/5/2.5/10/25/100/250/500/1,000 v. D.C. 0/2.5/10/25/100/250/500/1,000 v. A.C. 0/50  $\mu$ A/5/50/500 mA. 12 amp. D.C. 0/60K/6 meg. 60 meg  $\Omega$ . £8.87, Post paid.**



**MODEL TE-90 50,000 O.P.V. Mirror scale, overload protection. 0/3/15/60/300/600/1,200 v. D.C. 0/6/30/120/300/1,200 v. A.C. 0/3/60/600 MA. D.C. 16K/160K/1.6/16 MEG. -20 to + 63 dB. £7.50. P. & P. 15p.**



**TMK MODEL TW-20CB FEATURES RESETTING OVERLOAD BUTTON. Sensitivity: 20K  $\Omega$ /Volt D.C. 5K  $\Omega$ /Volt A.C. D.C. Volts: 0-0.5, 2.5, 10, 50, 250, 1,000V. A.C. Volts: 0-2.5, 10, 50, 250, 1,000V. D.C. Currents: 0-0.05, 0.5, 5, 50, 500mA. -10 amp. Resistance: 0-5K, 50K, 0-500K. 5 MEG. Decibels: -20 to + 52db. £11.50. P. & P. 17 1/2p.**



**MODEL AS-100D. 100K  $\Omega$ /Volt 5 in. mirror scale. Built-in meter protection 0/3/12/60/120/300/600 1,200 v. D.C. 0/6/30/120/300/600 v. A.C. 0/10  $\mu$ A/6/60/300mA/12 Amp. 0/2K/200K/2000M. -20 to + 17 dB. £12.50. P. & P. 17 1/2p.**



**TMK LAB TESTER 100,000 O.P.V. 6 1/2 in. Scale Buzzer Short Circuit Check. Sensitivity: 100,000 OPV D.C. 5  $\Omega$ /Volt A.C. D.C. Volts: 5, 2.5, 10, 50, 250, 1,000V. A.C. Volts: 3, 10, 50, 250, 500, 1,000V. D.C. Current: 10, 100  $\mu$ A, 10, 100, 500mA, 2.5, 10 amp. Resistance: 1K, 10K, 100K, 10MEG, 100MEG. Decibels: -10 to +49 db. Plastic Case with carrying handle. Size 7 1/2 x 6 1/2 x 3 1/2. £18.90. P. & P. 25p.**



**SKYWOOD SW-500 50 K  $\Omega$ /Volt. Mirror scale D.C. volts: 0.6/3/12/30/300/600. A.C. volts: 3/30/300/600. D.C. current: 20  $\mu$ A/6/60/600mA. Resistance: 10K/100K/1 Meg. Decibels: -20 to + 57 dB. £7.50. P. & P. 15p.**



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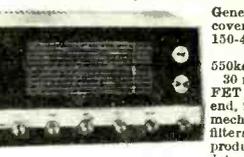
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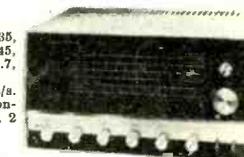


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General coverage 150-400 kc/s, 550kc/s-30 mc/s. FET front end, 2 mech. filters, product detector.



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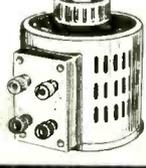


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BA115	8p	MPI105	40p	OC35	50p	2N1711	25p
BC107	12p	NKT124	30p	OC36	63p	2N2147	82p
BC108	12p	NKT125	40p	OC41	25p	2N2148	63p
BC109	12p	NKT126	37p	OC42	30p	2N2160	62p
BC147	15p	NKT128	25p	OC44	15p	2N2368	17p
BC148	15p	NKT135	26p	OC45	15p	2N2369	17p
BC149	15p	NKT137	32p	OC71	15p	2N2369A	20p
BC158	17p	NKT210	25p	OC72	23p	2N2646	50p
BC169C	19p	NKT211	25p	OC73	23p	2N2904	44p
BC182	12p	NKT212	25p	OC76	25p	2N2904A	44p
BC182L	10p	NKT213	25p	OC77	40p	2N2905	65p
BC183	9p	NKT214	23p	OC81	23p	2N2905A	75p
BC183L	9p	NKT215	21p	OC81D	20p	2N2906	46p
BC184	15p	NKT216	46p	OC81Z	25p	2N2906A	54p
BC184L	15p	NKT217	50p	OC82	25p	2N2926 all	10p
BC212	17p	NKT218	25p	OC82D	15p	colours	10p
BC212L	17p	NKT219	25p	OC83	23p	2N3053	25p
BCY30	25p	NKT223	27p	OC84	25p	2N3054	63p
BCY31	48p	NKT224	25p	OC139	25p	2N3055	75p
BCY32	50p	NKT225	21p	OC140	35p	2N3702	11p
BCY33	20p	NKT229	29p	OC170	25p	2N3703	10p
BCY34	25p	NKT230	31p	OC171	30p	2N3704	11p
BCY38	20p	NKT238	19p	OC200	37p	2N3705	10p
BCY70	17p	NKT239	23p	OC201	47p	2N3706	9p
BCY71	17p	NKT240	20p	OC202	63p	2N3707	11p
BCY72	16p	NKT241	21p	OC203	37p	2N3708	7p
BD121	£1.10	NKT242	15p	OC204	40p	2N3709	9p
BD123	£1.10	NKT243	56p	OC205	65p	2N3710	9p
BD124	£1.03	NKT244	17p	OC206	75p	2N3711	9p
BDY20	£1.05	NKT245	17p	OC207	75p	2N3819	35p
BF115	25p	NKT261	21p	OC207	75p	2N3820	60p
BF163	40p	NKT262	19p	OC207	75p	2N3826	30p
BF167	25p	NKT264	21p	OC207	75p	2N4058	17p
BF173	30p	NKT271	18p	OC207	75p	2N4060	20p
BF178	30p	NKT272	18p	OC207	75p	2N4061	20p
BF180	37p	NKT274	18p	OC207	75p	2N4062	20p
BF181	37p	NKT275	23p	OC207	75p	2N4284	15p
BF184	25p	NKT279A	12p	OC207	75p	2N4287	15p
BF185	25p	NKT281	29p	OC207	75p	2N4289	15p
BF194	17p	NKT302	87p	OC207	75p	2N4871	40p
BF195	15p	NKT304	75p	OC207	75p	3N84	£1.30
BF196	15p	NKT351	75p	OC207	75p	3N128	69p
BF200	35p	NKT401	71p	OC207	75p	3N140	76p
BFX13	25p	NKT402	77p	OC207	75p	3N141	73p
BFX29	31p	NKT403	65p	OC207	75p	3N152	86p
BFX84	26p	NKT404	60p	OC207	75p	3N152	86p
BFX85	34p	NKT405	75p	OC207	75p	40250	25p
BFX86	25p	NKT406	62p	OC207	75p	40309	15p
BFX87	30p	NKT420	£1.83	OC207	75p	40310	45p
BFX88	25p	NKT451	58p	OC207	75p	40312	48p
BFY50	23p	NKT452	54p	OC207	75p	40320	36p
BFY51	19p	NKT453	50p	OC207	75p	40360	40p
BFY52	20p	NKT603F	30p	OC207	75p	40361	48p
BFY53	16p	NKT613F	30p	OC207	75p	40362	58p
BFY90	67p	NKT674F	30p	OC207	75p	40406	56p
BSX19	16p	NKT676F	30p	OC207	75p	40407	39p
BSX20	16p	NKT677F	28p	OC207	75p	40408	51p
BSX21	37p	NKT713	29p	OC207	75p	40409	54p
BSY27	20p	NKT717	44p	OC207	75p	40409	54p
BSY29	25p	NKT734	25p	OC207	75p	40468A	35p
BSY95A	15p	NKT736	32p	OC207	75p	40600	50p
BY100	20p	NKT773	25p	OC207	75p	40601	50p
BYX10	15p	NKT781	29p	OC207	75p	40602	40p
BYZ10	40p	NKT10339	25p	OC207	75p	40602	40p
BYZ12	30p	NKT10419	19p	OC207	75p	40603	49p

<b>TRIACS</b>			
2N5756	2.5 Amp (RMS) 400 PIV TO-5 Mod.	95p	
40486	6 Amp (RMS) 400 PIV TO-5 Mod.	£1.20	
40430	6 Amp (RMS) 400 PIV TO-66	£1.01	
40432	6 Amp (RMS) 400 PIV TO-66	£1.50	
40512	2.2 Amp (RMS) 400 PIV TO-5 Mod.	£1.45	
	* these types have integral triggering		
40576	15 Amp (RMS) 400 PIV TO-66	£1.70	
SCI146B	10 Amp 200 PIV Plastic Flat-pack	£1.25	
SCI146D	10 Amp 400 PIV Plastic Flat-pack	£1.75	
ST2	Bi-lateral avalanche trigger diode	47p	

<b>THYRISTORS</b>			
CR1/051C	1 Amp 50 PIV TO-5	40p	
CR1/401C	1 Amp 400 PIV TO-5	50p	
2N3525	5 Amp 400 PIV TO-66	£1.09	
40739	10 Amp 400 PIV Stud Mounting	£1.65	

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1 and 1/2 watt 5% Each 2p  
Packs of 10 (of one value/wattage) Per pack 15p

**PRESETS—P.C. Type 0.3 watt**  
Standard size 5p  
Sub-miniature 7p  
(Available vertical or horizontal mounting.) Usual values 100 ohms to 5 Meg.

**POTENTIOMETERS**  
Log or Lin less switch 17p  
Log or Lin DP switch 27p  
Log or Lin Stereo L/S 50p  
Values: 5K, 10K, 25K, 50K, 100K, 250K, 500K, 1 Meg, 2 Meg.

**CAPACITORS—Mullard Miniature Electrolytic C426 series**

Mfd.	Volt.	Wkg.	Price
2.5	16	8p	
10	16	6p	
20	16	6p	
40	16	6p	
80	16	6p	
1-6	25	8p	
6-4	25	6p	
12-5	25	6p	
25	25	6p	
50	25	6p	
80	25	6p	
4	40	8p	
6	40	6p	
12	40	6p	
36	40	6p	
16	40	6p	
50	40	6p	

**Mullard Metallised Polyester C280 series**

Mfd.	Volt.	Wkg.	Price
0.01	16	3p	
0.022	16	3p	
0.033	16	3p	
0.047	16	4p	
0.068	16	4p	
0.15	16	4p	
0.22	16	5p	
0.33	16	7p	
0.47	16	8p	
0.68	16	11p	
1.0	16	14p	
1.5	16	20p	
2.2	16	24p	

**Mullard Electrolytic C437 series**

Mfd.	Volt.	Wkg.	Price
250	16	9p	
300	16	12p	
400	16	16p	
640	16	18p	
1,000	16	18p	
1,600	16	18p	
2,500	16	18p	
400	25	12p	
640	25	15p	
100	40	18p	
160	40	12p	
250	40	15p	
400	40	18p	

**Mullard Sub-Miniature Ceramic Plate C333 series**  
63 volt working. Range 1-8pf to 220pf (usual pref. values).  
Packs of 6 (any values) 30p

**NEONS**  
Miniature neon bulbs 0.6mA 65v. AC, 90v. DC.  
Pack of 5 for 30p

Panel neon indicators: main voltage, Red lens—round square or arrow shaped faces 20p

**VEROBOARD**  
2.5" x 17" x 0.15" 57p  
2.5" x 5" x 0.15" 23p  
2.5" x 3.75" x 0.15" 19p  
3.75" x 17" x 0.15" 79p  
3.75" x 5" x 0.15" 30p  
3.75" x 3.75" x 0.15" 22p  
2.5" x 5" x 0.1" 25p  
2.5" x 3.75" x 0.1" 23p  
Spot face cutters 30p  
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TO-5 (clip-on) Pack of 4 for FINNED type for 2xTO-3 ready drilled at 43p  
FINNED type undrilled for plastic power at 34p

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Designers Guide to British Transistors (data book) £1.25  
R.C.A. Hobby circuits manual 110 Semiconductor Projects (Marston) £1.25  
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SN7404N	Hex Inverter	32p	27p	22p
SN7410N	Triple			

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0A2	0-30	6BW6	0-72	6V6GT	0-33	20D1	0-65	150B2	0-58	DL33	0-35	ECH81	0-29	
0B2	0-30	6BW7	0-65	6X4	0-22	20D4	1-02	150C2	0-30	DL92	0-29	ECH83	0-40	
0C4	0-23	6C4	0-25	6X5GT	0-22	20F2	0-70	301	1-00	DL94	0-32	ECH84	0-38	
1A3	0-23	6C6	0-19	6Y6G	0-43	20L1	0-98	302	0-82	DL94	0-37	ECL80	0-35	
1A5	0-25	6C9	0-73	6Y6G	0-83	20P1	0-88	303	0-75	DM70	0-30	ECL82	0-32	
1A7GT	0-37	6C12	0-29	7B7	0-68	20P3	0-90	305	0-83	DM71	0-38	ECL83	0-62	
1D5	0-38	6C17	0-23	7B7	0-35	20P4	0-93	306	0-65	DM74/350	0-38	ECL84	0-30	
1D6	0-48	6CD6G	1-15	7C6	0-30	20P5	1-00	807	0-59	EW49	0-38	EOL85	0-65	
1FD1	0-35	6CH6	0-38	7F8	0-63	25A6G	0-29	956	0-10	DY86/70-29	0-38	ECL86	0-40	
1FD9	0-22	6CL6	0-43	7H7	0-28	25B5	0-29	1821	0-53	DY86/20-48	0-38	EM21	0-43	
1G6	0-30	6CW4	0-68	7E7	0-28	25Y5	0-38	8743	0-50	E80F	1-20	EF36	0-33	
1HG0T	0-35	6D3	0-38	7V7	0-25	25Z4	0-43	6060	0-30	E83F	1-20	EF37A	0-35	
1L4	0-13	6D6	0-15	9BW6	0-60	25Y6	0-30	7193	0-53	E88CC	0-60	EF39	0-40	
1LD5	0-30	6F1	0-63	9D7	0-78	25Z5	0-40	7475	0-70	E180F	0-96	EF40	0-50	
1LN5	0-40	6P6	0-63	10C1	1-25	25Z0G	0-43	A1834	1-00	E182CC/1-13	0-50	EF41	0-50	
1N5GT	0-39	6F6G	0-43	7H7	0-30	30C1	0-30	A2134	0-98	E1148	0-53	EF42	0-33	
1R5	0-28	6F8	0-17	10C14	0-33	30C15	0-65	3042	0-75	EA50	0-18	EF54	0-98	
1R4	0-24	6F13	0-33	10D1	0-50	30C17	0-80	AC044	1-18	EA76	0-88	EF73	0-33	
1R5	0-22	6F14	0-75	10F1	0-75	30C18	0-84	AC2PEN	0-98	EAC80	0-33	EF80	0-23	
1U4	0-29	6F15	0-65	10P9	0-45	30C15	0-80	AC2PEN	0-98	EAC91	0-38	EF89	0-48	
1U5	0-48	6F18	0-45	10P18	0-35	30FL1	0-64	0-78	0-78	EAF42	0-50	EF85	0-29	
2D21	0-35	6F23	0-72	10L14	0-37	30FL2	0-75	AC2PEN	0-98	EAC91	0-38	EF89	0-48	
3A4	0-20	6F24	0-85	10L1D11	0-63	30FL12	0-80	AC2PEN	0-98	EB41	0-50	EF89	0-25	
3A5	0-10	6F25	0-85	10P12	0-65	30FL14	0-73	0-98	0-98	EB91	0-12	EF91	0-17	
3B7	0-25	6F28	0-29	10P13	0-65	30L1	0-32	AC/TH1	0-50	EBC41	0-48	EF92	0-13	
3D6	0-19	6F28	0-70	10P14	1-10	30L15	0-64	AC/TP	0-98	EBC81	0-33	EF97	0-55	
3Q4	0-38	6A6G	0-75	10P18	0-33	30L17	0-78	A160	0-78	EBC90	0-20	EF98	0-66	
355GT	0-35	6H6GT	0-15	12A6	0-63	30P4MR	0-8	ARF3	0-35	EB091	0-30	EF83	0-30	
3B4	0-23	6K3	0-19	12A6C	0-40	30P12	0-69	ATP4	0-12	EBF80	0-40	EF184	0-80	
3V4	0-22	6J6	0-18	12AD6	0-40	30P16	0-33	AZ1	0-40	EBF83	0-40	EF184	0-80	
5R4GY	0-53	6J7G	0-24	12A6E	0-48	30P18	0-33	AZ31	0-48	EBF89	0-32	EH90	0-38	
5V4G	0-38	6J7GT	0-38	12A76	0-23	30P19/30P4	AZ41	0-53	EEL21	0-60	EK90	0-24		
5Y3GT	0-28	6K7G	0-10	12A7T	0-19	0-60	0-60	B319	0-32	EC53	0-63	EL32	0-18	
5Z3	0-45	6K7GT	0-23	12A08	0-24	30PL1	0-69	CL33	0-98	EC54	0-50	EL34	0-53	
5Z4G	0-25	6K8G	0-20	12A07	0-23	30PL12	0-37	CV6	0-53	EC70	0-24	EL37	0-87	
6/30L2	0-58	6L1	0-98	12AV6	0-28	30PL13	0-78	CV888	1-10	EC86	0-83	EL41	0-45	
6A8G	0-33	6L6GT	0-38	12AX7	0-23	30PL14	0-75	CY1C	0-53	EC86	0-80	EL42	0-53	
6A7C	0-15	6L7GT	0-63	12AX7	0-78	30PL15	0-98	CY31	0-38	EC92	0-35	EL81	0-80	
6A45	0-25	6L18	0-45	12BA6	0-30	35A3	0-50	D63	0-25	ECC32	1-58	EL84	0-38	
6A45	0-25	6L19	1-38	12B6E	0-30	35A5	0-75	D77	0-12	ECC33	1-58	EL84	0-38	
6A46	0-30	6L20	0-48	12B7H	0-40	50B5	0-70	DP21	0-14	ECC86	0-40	EM80	0-30	
6A45	0-12	6N7GT	0-40	12E1	0-25	33L6GT	0-44	DAF91	0-25	ECC81	0-19	EL86	0-40	
6A44	0-83	6P15	0-24	12J7GT	0-33	35W4	0-23	DAF96	0-35	ECC82	0-23	EL91	0-23	
6A46	0-17	6P28	1-25	12K7G	0-30	35Z3	0-50	DOC90	1-00	ECC83	0-23	EL95	0-35	
6A45	0-28	6Q7G	0-30	12K7GT	0-34	35Z4GT	0-24	DD4	0-53	ECC84	0-30	EM34	0-80	
6A46	0-10	6Q7GT	0-43	12Q7GT	0-28	35Z5GT	0-30	DF33	0-39	ECC85	0-28	EM80	0-38	
6A76	0-20	6R7G	0-35	128A7GT	0-40	50B5	0-70	DP21	0-14	ECC86	0-40	EM81	0-42	
6A76	0-25	6R7	0-55	0-40	50C5	0-32	DP96	0-35	ECC88	0-35	EM84	0-34		
6A76	0-30	68A7GT	0-35	50C6G2	0-17	DF97	0-63	ECC189	0-48	EM87	0-38	N339	1-25	
6B8G	0-13	68CTGT	0-33	50L6GT	0-45	DH63	0-30	ECC80	0-58	EY31	0-37	N359	0-48	
6BA6	0-23	68G7	0-33	128H7	0-15	72	0-33	DH76	0-28	ECC80	0-58	EY31	0-37	
6B26	0-24	68H7	0-53	128J7	0-23	77	0-53	DH77	0-20	ECC80	0-58	EY31	0-37	
6B16	0-43	68GT	0-35	128K7	0-24	75A2	0-43	DH81	0-58	ECF82	0-20	EY83	0-55	
6B76	0-48	68K7GT	0-23	128Q7GT	0-23	85A3	0-40	DH101	1-25	ECF86	0-65	EY86/7	0-50	
6B25	0-24	68N7GT	0-23	0-50	90A6	3-38	DK32	0-37	ECF84	EY88	0-43	PAB80	0-53	
6BQ7A	0-38	68Q7	0-33	14H7	0-48	90AV	3-38	DK40	0-55	2-10	EY91	0-83	PC86	0-52
6B77	0-79	6U4GT	0-80	18	0-63	90C9	1-70	DK91	0-28	ECH21	0-63	EZ35	0-25	
6B83	0-63	6U7G	0-55	19AQ5	0-24	90CV	1-63	CK92	0-43	ECH35	0-29	EZ40	0-40	
6B87	1-25	6V6G	0-18	19H1	2-00	90C1	0-80	DK96	0-37	ECH42	0-64	EZ41	0-43	

EZ80	0-23	PC900	0-38	PY81	0-27	UY85	0-29	and diodes	AF139	0-65	GD4	0-32	OC22	0-38	
EZ81	0-24	PC84	0-32	PY82	0-27	U10	0-45	1N124	0-53	AF178	0-68	GD5	0-28	OC23	0-38
EZ90	0-22	PC85	0-33	PY83	0-29	U12/14	0-38	2N404	0-18	AF180	0-48	GD6	0-28	OC24	0-38
FW4	/500	PC88	0-48	PY88	0-34	U16	0-75	2N966	0-53	AF181	0-70	GD8	0-20	OC28	0-38
FW4/800	0-75	PC89	0-48	PY91	0-63	U17	0-35	2N1766	0-50	AF182	0-45	GD8	0-20	OC29	0-60
GZ30	0-35	PC806	0-78	PY800	0-38	U19	1-73	2N2297	0-83	AF239	0-38	GD10	0-20	OC28	0-60
GZ32	0-45	PC800	0-64	PY800	0-38	U19	1-73	2N2297	0-83	AF239	0-38	GD11	0-20	OC29	0-60
GZ33	0-70	PC800	0-40	QV03/10	U26	0-59	2N2613	0-38	BI181	0-50	GD15	0-40	OC38	0-43	
GZ34	0-75	PC806	0-33	PCF82	0-23	U21	0-35	2N1766	0-50	AF182	0-45	GD8	0-20	OC29	0-60
GZ35	0-75	PC806	0-40	QV75/20	U33	1-48	2N1766	0-50	AF182	0-45	GD8	0-20	OC29	0-60	
HAB80	0-45	PC806	0-50	Q8150/15	U35	0-83	2N3703	0-19	BA116	0-25	GET113	0-20	OC43	1-18	
HL13C	0-20	PCF87	0-80	0-63	U37	1-76	2N3709	0-20	BA129	0-13	GET116	0-40	OC44	1-10	
HL23D	0-40	PCF200	0-67	QV04/7	0-63	U45	0-75	2N3866	0-10	BA130	0-10	GET118	0-20	OC45	1-13
HL41D	0-88	PCF800	0-65	R10	0-75	U47	0-68	2N3988	0-50	BCY10	0-45	GET119	0-20	OC46	1-18
HL50	0-35	PCF801	0-35	R11	0-98	U49	0-59	2N825	0-50	BCY12	0-50	GET123	0-38	OC58	1-13
HN309	1-37	PCF802	0-45	R16	1-75	U50	0-28	AA119	0-15	BCY33	0-20	GET187	0-43	OC70	1-13
HVR2A	0-53	PCF805	0-64	R17	0-88	U76	0-24	AA120	0-15	BCY34	0-23	GET187	0-43	OC71	1-13
HVR2A	0-53	PCF806	0-64	R18	0-50	U78	0-22	AA129	0-15	BCY38	0-23	GET187	0-43	OC72	1-13
HW3	0-38	PCF808	0-73	R19	0-33	U107	0-92	AA133	0-18	BCY39	0-25	GET188	0-50	OC74	0-83
IW4/350	0-38	PCF810	0-76	R20	0-59	U153	0-27	AO107	0-15	BC107	0-13	GET188	0-50	OC75	1-13
IW4/500	0-38	PCF820	0-82	R25	0-38	U191	0-65	AO113	0-25	BC108	0-13	GET189	0-23	OC76	1-18
IW4/500	0-38	PCF82	0-37	RG1/240	0-40	U192	0-27	AO114	0-40	BC113	0-25	GET190	0-23	OC77	1-27
KT2	0-25	PC83	0-84	R30	1-98	U193	0-34	AO127	0-20	BC115	0-15	GET190	0-23	OC78	1-15
KT2	0-25	PC84	0-38	RK34	0-38	U251	0-73	AO128	0-20	BC116	0-25	GET197	0-23	OC78D	0-15
KT8	1-73	PC86	0-43	8P13C	0-83	U281	0-40	AO154	0-25	BC118	0-23	GET198	0-23	OC79	0-40
KT14	1-03	PC88	0-76	8P42	0-76	U282	0-40	AO166	0-20	BC211	0-38	GET13	0-18	OC81	1-13
KT14	1-03	PC800	0-76	8P61	0-33	U297	0-50	AO176	0-25	BD119	0-45	GET36	0-23	OC82	0-45
KT63	0-25	PC801	0-98	THAB	0-80	U301	0-53	AO185	0-25	BF154	0-25	GET36	0-50	OC82	0-13
KT66	0-63	PC805	0-48	TH233	0-98	U329	0-73	AO186	0-25	BF159	0-25	GET45	0-33	OC82D	0-15
KT74	0-63	PC815	0-45	TH2620	0-98	U301	0-29	AO167	0-60	BF163	0-20	GET56	0-75	OC83	0-20
KT76	0-63	PC800	0-33	UA030	0-33	U403	0-33	AO168	0-38	BF173</					

# AUDIOTRINE A55 HIGH QUALITY STEREO SYSTEM

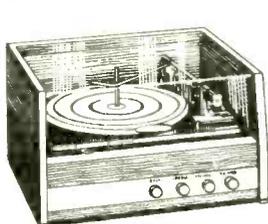
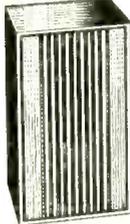
### 5 + 5 WATT OUTPUT

#### Garrard 5200 Changer

with low mass pick-up arm and Stereo Cartridge. Controls: TREBLE, BASS, VOLUME STEREO BALANCE.

Operation on 200-250 v. A.C. mains. Output rating I.H.F.M.

Luxurious Teak Veneer Finished Cabinets. Transparent plastic (tinted) cover included for main unit. Silver finished fascia plate and matching control knobs.



### PAIR OF LOUDSPEAKER UNITS

incorporating high flux 8in. x 5in. speaker. Size approx. 13 x 7 1/2 x 8 1/2 ins.

### PRICE COMPLETE ONLY

Carr. £1.25  
Terms: Deposit £5.50 and 9 monthly payments £4.50 (Total £46)

# FANE 807 HIGH FIDELITY LOUDSPEAKER

A full range 8in. 10 watt speaker. Unit for excellent sound quality in suitable enclosure. Roll P.V.C. cone surround and long throw voice coil to achieve very low fundamental resonance at 30 c.p.s. Tweeter cone is fitted to extend high note response. Frequency range 20-15KHz. Impedance 3Ω or 8-15Ω. Cast Chassis. REMARKABLE VALUE AT ONLY **£3.50**

### AUDIOTRINE HIGH FIDELITY LOUDSPEAKERS

Heavy construction. Latest high efficiency ceramic magnets. Treated Cone surround or "L" indicates Roll Rubber surround. "D" indicates Tweeter Cone providing extended frequency range up to 15,000 c.p.s. Exceptional performance at low cost. Impedance 3 or 8-15 ohms.

WHEN ORDERING PLEASE STATE IMPEDANCE			
HF 801D 8" SW	£2.71	HF 120D 12" 15W	£4.49
HF 102D 10" 10W	£3.40	HF 128 12" 15W	£5.25
HF 120 12" 15W	£3.99	HF 128D 12" 15W	£5.75

# FANE ULTRA HIGH POWER LOUDSPEAKERS

All power ratings are R.M.S. continuous. 2 years' guarantee. High flux ceramic magnets. Heavy cast chassis. ALL CARRIAGE FREE.

'POP' 100	'POP' 60	'POP' 50
18in. 100 watt 14,000 gauss 8/15Ω	15in. 60 watt 14,000 gauss 8/15Ω	12in. 50 watt 13,000 gauss 15Ω
<b>£22.05</b>	<b>£12.90</b>	<b>£10.50</b>

Dep.: £6 and 9 monthly payments £2 (Total £24).  
Dep.: £3.30 and 9 monthly payments £1.30 (Total £15).  
Dep.: £2 and 9 monthly payments £1.15 (Total £12.35).  
PAIR SUITABLE ALL PURPOSES.  
FOR BASS GUITAR OR ELECTRONIC ORGAN, ETC.

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Dual cone 15Ω (for use other than Bass Guitar or Electronic Organ). Carr. free. Output for 3-15 ohm speakers. Max. sensitivity 5mV. Output rating I.H.F.M. In fully enclosed enamelled case, approx. 9 1/2 x 2 1/2 x 6 1/2 in. Attractively brushed silver finish fascia plate 10 1/2 x 4 1/2 in. and matching knobs. Complete kit of parts with full wiring diagrams and instructions. Carr. 40p. **£6.75**

# R.S.C. TA6 6 Watt High Fidelity Solid State Amplifier

200-250v. A.C. mains operated Frequency Response 30-20,000 c.p.s. -2dB. Harmonic Distortion 0.3% at 1,000 c.p.s. Separate Bass and Treble controls. 3 input sockets for Mike, Gram, Radio or Tape. Input selector switch. Output for 3-15 ohm speakers. Max. sensitivity 5mV. Output rating I.H.F.M. In fully enclosed enamelled case, approx. 9 1/2 x 2 1/2 x 6 1/2 in. Attractively brushed silver finish fascia plate 10 1/2 x 4 1/2 in. and matching knobs. Complete kit of parts with full wiring diagrams and instructions. Carr. 40p. **£7.50**

# R.S.C. BATTERY/MAINS CONVERSION UNITS

Type BM1. An all-dry battery eliminator. Size 9 1/2 x 4 1/2 x 2 1/2 in. approx. Completely replaces batteries supplying of 1.5 v. and 90 v. where A.C. mains 200/250 v. 50 c/s. is available. Complete kit with diagram £3 or, **READY FOR USE £3.50.**

# HIGH QUALITY LOUDSPEAKERS

In teak or afornormosa veneered cabinets. L13 3 or 15 ohms. 13 x 8in. 8-10 Watt M o d e 1 Gauss 10,000 **£5.25** Carr. 40p  
L12 12in. 20 Watt Model. Gauss 11,000 lines. Size 18 x 18 x 10 in. approx. 15 watt/speakers. £27.50 Carr. 40p  
L11 12in. 20 Watt Model. Gauss 11,000 lines. Size 18 x 18 x 10 in. approx. 15 watt/speakers. £27.50 Carr. 40p  
L10 12in. 20 Watt Model. Gauss 11,000 lines. Size 18 x 18 x 10 in. approx. 15 watt/speakers. £27.50 Carr. 40p

# AUDIOTRINE HI-FI SPEAKER SYSTEMS

Consisting of matched 12in. 15,000 line 15 watt 16 ohm high quality speaker, cross-over unit and tweeter. Smooth response and extended frequency range ensure surprisingly realistic reproduction. Carr. 30p. **£5.75**

Or SENIOR 15 WATT inc. HF126 15,000 line Speaker £6.75. Carr. 35p.  
**HI-FI LOUDSPEAKER ENCLOSURES**  
Teak or Afornormosa veneer finish. Modern design. Acoustically lined. All sizes approx. Carr. 25p extra.  
JES Size 16 x 11 x 9 in. Pressurised. Gives pleasing results with any 8in. HI-FI speaker. **£4.75**  
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# TA12 MK III 6+5 + 6+5 WATT STEREO AMPLIFIER

FULLY TRANSISTORISED. 80WATT CONSTRUCTION HIGH FIDELITY OUTPUT OF 6.5 WATTS PER CHANNEL. Designed for optimum performance with any crystal or ceramic Gram P.U. cartridge, Radio tuner, Tape recorder, 'Mike' etc. 3 separate switched input sockets on each channel. \* Separate Bass and Treble controls \* Slide Switch for mono use \* Speaker Output 3-15 ohms \* For 200-250 v. A.C. mains \* Frequency Response 20-20,000 c.p.s. -2dB \* Harmonic Distortion 0.3% at 1000c.p.s. Hum and noise -70dB \* Sensitivities (1) 50 mV (2) 400 mV (3) 100 mV \* Handsome finish Facia Plate and Knobs. Output rating I.H.F.M. Complete kit of parts **£15.50** with full wiring diagrams and instructions. Carr. 40p.  
**FACTORY BUILT WITH 12 MONTH GUTEE. £19.50.** Or dep. £23 and 9 monthly pymts. £2.05 (Total £21.45). Or in Teak veneer housing. £23. Or Dep. £23 and 9 monthly pymts £2.55 (Total £25.95).

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F.W. Bridged 6/12v. D.C. Output Input Max. 18v A.C. 1a., 25p; 2a., 35p; 3a., 50p; 4a., 65p; 6a., 80p

# R.S.C. G66 6+6 WATT HIGH QUALITY STEREO AMPLIFIER

Individual/Ganged controls: Bass, Treble, Volume and Balance. Printed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F.M. Suitable for Crystal Pick-ups etc., and for loudspeaker output impedances of 3 to 15 ohms. For standard 200-250 v. A.C. mains operation. Attractive silver finished metal fascia plate and matching control knobs. Complete KIT of PARTS INCLUDING FULLY WIRED PRINTED CIRCUIT and comprehensive wiring diagram and instructions. **£9.99** No leaflet supplied for this unit. **£12.50.** Carr. 40p. Or Deposit £2 and 9 monthly payments of **£1.45** (Total **£15.05**).

# PACKAGE OFFER, SAVE APPROX. £4. Above G66 assembled in cabinet plus 2x POP 25/2 LOUDSPEAKER UNITS. £27.50 Carr. £1 (Total £28.50). Or Deposit £5.25 and 9 monthly payments £2.85 (Total £30.90).

# R.S.C. A10 30 WATT ULTRA LINEAR HI-FI AMPLIFIER

Highly sensitive. Push-Pull high output, with Pre-amp/Tone Control Stages. Hum level -70dB. Frequency response ±3dB 30-20,000 c/s. All high grade components. Valves 6F86, 6F86E, ECC83, 607, 607, GZ34. Separate Bass and Treble Controls. Sensitivity 36 millivolts. Suitable for High Impedance mic. or Outdoor Functions, etc. For use with Electronic Organ, Guitar, String Bass, etc. Gram, Radio or Tape. Reserve L.T. and H.T. for Radio Tuner. Two inputs with associated volume controls so that two separate inputs such as Gram and 'Mike' can be mixed. 200-250 v., A.C. For 3 & 15 Ω speakers. Complete Kit parts, wiring diagrams, instructions. **£15.75** Twin-handled perforated cover £1.75. Or factory built with EL24 output valves and 12 months' guarantee for £19.75. Tech. Bags apply to factory built units. Carr. 65p. **TERMS:** Deposit £4.00 and 9 monthly payments of £2.10 (Total £22.90). Send S.A.E. for leaflet.

# RSC BASS-REGENT 50 watt AMPLIFIER

A powerful high quality, all purpose unit. For lead, rhythm, bass guitar, and reliable components. FOUR JACK INPUTS and TWO VOLUME CONTROLS for simultaneous use of up to 4 pick-ups or 'mikes'. SEPARATE BASS AND TREBLE CONTROLS. OR SUPPLIED COMPLETE with matched twin loudspeaker unit as illustrated for £60. Carr. £1.50. Terms: Dep. £16 and 9 monthly payments £5.75 (Total £67.75). **ONLY £30** Carr. 90p

# THE 'YORK' HIGH FIDELITY 3'SPEAKER SYSTEM

Moderate size (approx. 25 x 14 x 10 in.). Range 30-20,000 c.p.s. Impedance 15 ohms. Performance comparable with units costing considerably more. Consists of (1) 12in. 15 watt Base unit with cast chassis. Roll rubber cone surround for ultra low resonance, and ceramic magnet. (2) 3-way quarter section series cross-over system. (3) 8 x 6in. high flux middle range speaker. (4) High efficiency tweeter. (5) Appropriate quantity acoustic damping material. (6) Teak veneered cabinet. (7) Circuit and full instructions. **£22.** Carr. 65p  
**REMARKABLE VALUE HEAR IT AT ANY BRANCH**

# R.S.C. SUPER 30 Mk II HIGH FIDELITY STEREO AMPLIFIER

**HIGH GRADE COMPONENTS. SPECIFICATIONS COMPARABLE WITH UNITS COSTING CONSIDERABLY MORE**  
Employing Twin Printed Circuits.  
200/250v. A.C. mains operation.  
**TRANSISTORS:** 9 high-quality types per channel.  
**OUTPUT (Per channel):** 10 Watts R.M.S. continuous into 16 Ω 15 Watts R.M.S. continuous into 3 Ω.  
**INPUT SENSITIVITIES:** Mag. P.U. 4 mV. Ceramic P.U. 35 mV. Tape Amp. 400 mV. A.V. 100 mV. Mic. 5 mV. Tape Head 2.5 mV.  
**FREQUENCY RESPONSE:** ±2dB. 10-20,000 c.p.s.  
**TREBLE CONTROL:** +17 dB to -14 dB at 10 Kc/s.  
**BASS CONTROL:** +17 dB to -15 dB at 50 c/s.  
**HUM LEVEL:** -80 dB.  
**HARMONIC DISTORTION:** 0.1% at 10 Watts 1,000 c.p.s.  
**CROSS TALK:** 52 dB at 1,000 c.p.s.

**EMINENTLY SUITABLE FOR USE WITH ANY MAKE OF PICK-UP OR MIC. (Ceramic or Magnetic, Moving Coil, Ribbon or Crystal) CURRENTLY AVAILABLE. SUPERB SOUND OUTPUT QUALITY CAN BE OBTAINED BY USE WITH FIRST-RATE ANCILLARY EQUIPMENT. COMPLETE KIT OF PARTS, point to point wiring diagrams & detailed instructions £23.25 Carr. 75p**  
**UNIT FACTORY BUILT £30.50**  
Or deposit £4 and 9 monthly payments £3.35 (Total £34.15).  
Or in Teak or Afornormosa veneer housing as illustrated. Carr. 75p.  
Terms: Deposit £4 and 9 monthly payments £3.75. (Total £37.75). Send S.A.E. for leaflet.

**CONTROLS:** 5-position Input Selector. Bass, Treble, Vol., Bal., Stereo/Mono Sw. Tape Monitor Sw., Mains Sw.  
**INPUT SOCKETS:** (1) P.U. (2) Tape Amp. (3) Radio. (4) Mic. or Tape Head. (Operation of Input Selector assures appropriate equalisation.)  
**CHASSIS:** Strong Steel construction. Approx. 12 x 8 x 8 in.  
**FACIA PLATE:** Attractive design in rigid "Plexiglas" with silver background. Spun silver matching control knobs as available.

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Superior Solid Natural Wood Construction. Record Playing units. Cut for Garrard 1025 2025 3000, AT60. SP25 etc. **£3.15**  
Available with Transparent plastic cover. **£6.30**

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**MONEY SAVING UNITS** Ready to Plug into Amplifier.  
**RP23C** Consisting of Garrard RP20 Mk. III fitted Goldring C890 high compliance ceramic Stereo/Mono cartridge with diamond stylus. Mounted on plinth. Transparent plastic cover included. Carr. 60p **£26.09**

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**OTHER TYPES** w/ P.U. Cartridges or 'Roll over' trans at lowest prices.

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BSY28	13p	OC35	25p
BSY29	13p	OC36	13p
BSA95A	13p	AD149	30p
OC41	13p	2N3055	65p
OC44	13p	2SO34	25p
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Complementary Set. NPN/PNP Germ. Trans. PAIR

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TRANSISTORISED FIELD RATEMETER type 1368A range 0.05 to 25 mr/hr in 5 ranges size 12 x 3 1/2 x 7 1/2 ins. £10 each. P. & P. 50p. SURVEY METER RADIAC No. 3. Hand portable size 9 1/2 x 5 x 5 1/2 ins. 3 ranges (scale changes) 0.03; 0.3; 3 R/H. Internal Ion Chamber. Nice condition £3 ea. P. & P. 50p. DOSIMETER 0-50R 0-150R and charger £2. P. & P. 7/6. Charger only 30/- P. & P. 33p.

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TRANSISTOR OSCILLATOR. Variable frequency 40 c/s to 5 kc/s. 5 volt square wave o/p, for 6 to 12v DC input. Size 1 1/2 x 1 1/2 x 1 1/2 in. Not encapsulated. Brand new. Boxed. 37p ea.

CRAMER TIMER 28V DC Sweep 1/100th sec & sweep 60 secs. 4" dial. Remote control stop/start reset £6.50.

RELAYS

G.E.C. Sealed Relays High Speed 24V. 2 make 2 break. 23p ea.

S.T.C. sealed 2 pole c/o, 2,500 ohms. (okay 24v) 13p ea; 12v 35p ea.

CARPENTERS polarised Single pole c/o 20 and 65 ohm coil as new, complete with base 37p ea. Single pole c/o 14 ohm coil 33p ea; Single pole c/o 45 ohm coil 33p ea.

POTENTIOMETERS

COLVERN Brand new. 50; 100; 250; 500 ohms; 1; 2.5; 5; 10; 25; 50k all at 13p ea. Special Brand new. MORGANITE 2.5K; 250K; 500K 2.5 meg. 1" sealed. 17p ea.

STANDARD 2 meg Log pots. Current type. 15p ea. INSTRUMENT 3" Colvern. 5; 25 ohms 35p ea.

BOURNE TRIM POTS. 10; 20; 50; 100; 200; 250; 500 ohms; 1; 2.5; 5; 25K at 35p ea.

ALMA precision resistors 100K; 400K; 497K; 998K; 1 meg—0.1. 27p ea.; 3-25K—0.1. 20p ea.

DALE heat sink resistors, non-inductive 50 watt. Brand new 8.2K at 13p ea.

MULLARD VINKORS. Brand new boxed. LA2104 60p ea.; LA2411 45p ea.; LA2503 30p ea.

SILVER ZINC Non-spill. Brand new. 7 1/2 V 5 cell. Size 1 1/2 x 1 x 1 1/2 oz. weight £1 ea. Single cell 1.5V 4AH size 1 1/2 x 1 x 3 1/2. 4oz. weight £1 ea.

MALLORY CELLS. 25p per set of 5.

CAPACITORS

ERIE feed through ceramic 2200 pf—4p ea. Sub-min. TRIMMER 1 square. 8, 5pf. Brand new 13p ea. Concentric TRIMMER 3/30 pf. Brand new 7p ea.

ELECTROLYTICS. Brand new. 250 mfd. 70V 23p ea. E.H.T. 2 mfd 5 KV. Brand new £1.50 ea. E.H.T. 0.1 mfd 7 KV at 40p ea.; 0.1 mfd 5 kv at 35p ea.

DECADE DIAL UP SWITCH. Finger-tip. Engraved 0/9. Gold plated contacts. Size 2 1/2" high, 2 1/2" deep 1/2" wide. 75p ea. Bank of 4 with escutcheon plates, etc. 2 1/2" high, 2 1/2" deep, 2 1/2" wide £2.50.

PHOTOCCELL equivalent OCP 71 13p ea. Photo-resist type Clare 703. (TO5 Case). Two for 50p.

BURGESS Micro Switches V3 5930. Brand new 13p ea. HONEYWELL. Sub-min. Microswitches type 11SM3-T. Brand new. 17p ea.

PANEL mounting lamp holders. Red. 9p ea. BRAND NEW PLUGS AND SOCKETS CANNON. 50 way DDM50 75p ea.; DDM50S 50p ea. £1 per pair.

As above but 25 way 50p ea. plug; 35p ea. socket; 75p per pair; 9 way 33p ea. plug and socket, 50p per pair. U.H.F. Plugs fit UR57, 59, 65 etc., 40p ea.

B.N.C. to U.H.F. Adaptor £1.37 ea.; Min. B.N.C. to U.H.F. £1.50 ea.; "T" Junction B.N.C. £1 ea.; B.N.C. plug to B.N.C. plug £1 ea.; B.N.C. Right angle £1 ea.; Min. B.N.C. right angle £1.25 ea.; Min. socket round 50p ea. Standard B.N.C. round 35p ea. Many others too numerous to list. All prices quoted for 'one off'.

TRANSFORMERS

All standard inputs. STEP DOWN ISOLATING trans. Standard 240V AC to 120V tapped 60-0-60 700W. Brand new. £5 ea. Transformer 0-215-250 120 MA.; 6.3V 4A CT x 2; 2 x 6.3v 0.5A and separate 90v 100 MA £1.25 ea. P. & P. 20p. Matching contact cooled bridge rectifier 37p ea. 4.5V 40 amp (180Va) £1.75 ea. incl. postage or 3 for £4.50 incl. postage. Designed to be Series paralleled. Farneco 6.3v 2 amp x 4—£1.13 ea. Gard/Parin/Part. 450-400-0-400-450, 180 MA. 2 x 6.3v. £3 ea.

CHOKES. 5H; 10H; 15H; up to 120mA, 42p ea. Up to 250mA 63p ea. Large quantity LT, HT, EHT transformers. Your requirements, please.

GROUND PLANE ANTENNA. Ex-admiralty. Brand new boxed. Adjustable 90-160 megs. (Like umbrella) £12.50. Carr. £1.

NUCLEONIC INSTRUMENTS

Pulse analyser N101; Scaler 1009E; Coincidence unit 1038C; Anti coincidence unit Panax AU480; Amplifier N567; A/B/G Radiation Monitor 1257A; complete 1339A system A/B/G; EHT Potentiometer unit 1007; 1430 amplifier CF and head; Some scintillation castles; radiation monitor 1320C and 1320X (X-ray); survey meters no. 2 and 3; Rate-meter scintillation 1368A; Fast neutron 1262C; Fluorimeter 1080A and many others. Also 2000 SERIES. Amp 2002A; Low level amp 2024; PU's 2004; 2005B; nanosec time amplitude converter 2011A; pulse amplitude analyser 2010B; discriminator 2007B; high level amp 2025 and others. Information available.

RACAL RAI7K receivers £250. Racal RA98A Automatic SSB adaptor for above. Brand new crated. £75. MARCONI TF1370A Wide Range R.C. Oscillator. As new Current model £200.

TEST GEAR

OSCILLOSCOPES

- E.M.I. WM16 DB—24 megs each channel. £175 only. E.M.I. WM 2 DC—13 mc/s £25. E.M.I. WM 8 DC—15 mc/s £40. SOLARTRON CD1015 DC—20 megs. £55. SOLARTRON 7115.2 D.B. DC—9 mc/s. In fine condition £50. SOLARTRON 843 DC—15 mc/s Brand new £85 Good condition £50. SOLARTRON DC—10 mc/s. CD513—£35, 513.2—£40. CD5238—£45. SOLARTRON CT316 (D300 range) DC—6 megs. £17.50. SOLARTRON Storage scope QD910 £150. COSSOR 1049 Mk. 3. DB. £25. HARTLEY 13A DB. £25. CT52 Min. scope. £17.50. All carefully checked and tested. Carriage £1.50 extra.

MARCONI

- TF 1152 Power Meter. New P.O.R. TF 1026 Frequency Meter £12.50. Carr. 75p. TF 329 Magnification Meter. As new condition £60. TF 195 Audio Generator £10. Carr. £1.50. TF 801A Signal generator £35. Carr. £1.50. TF 806 Magnification Meter £45. Carr. £1. TF 369 N. 5 Impedance Bridge £70. Carr. £1.50. TF 144G Signal Generator. Serviceable. Clean £15. In exceptional condition £25. Carr. £1.50. TF 885 Video Oscillator Sine/Square £35 Carr. £1.50. TF895/1 £55. Carr. £1.54. TF 1343/2 'X' Band gen. £35. Carr. £1.50.

SOLARTRON

- Laboratory amplifier AWS51A. 15c/s—350ke/s £35 Carr. £1. Stabilised P.U. SRS 151A £20. Carr. £1.50. Stabilised P.U. SRS 152 £15. Carr. £1.50. Precision Millivoltmeter VP252. £25. Carr. £1. Process Response Analyser. Fine condition £250 Oscillator type OS 101. £30. Carr. £1.50. D.C. Amplifier type AA900. £30. Carr. £1. Storage Oscilloscopes QD910. £150. Carr. £1.50.

AVO

- Testmeter No. 1 £12 ea. Carr. 75p. Electronic Testmeter CT 38. Complete £18 Carr. £1.

CINTEL

- Sine and Pulse Generator type 1873 £15. Carr. 75p.

AIRMEC

- Signal Generator type 701. £25. Carr. £1.50.

MARCONI TF 1277. Colour studio scope, will line select. In superb condition. £120.

LIMITED QUANTITY

TELEQUIPMENT D43R. Brand new with TD41 TB. £80 with 15 mc/s amp. £105.

BRADLEY ATTENUATORS 0/500 meg cycles. 0/12 db and 0/120 db—£20 per pair.

HEWLETT PACKARD. Attenuators 0/500 meg cycles. 0-132 db. 1 db steps. £40.

BECKMAN MODEL A. Ten turn pot complete with dial. 100K 3% Tol 0.25%—only £2.13 ea.

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BRUEL & KJOER Automatic Vibration Exciter Control type 1018. £140.

DVM's BIE 2114 £50 ea.; BIE 2116 £50 ea. Carr. £1.50.

AMERICAN TRIPLETT Generators type 1632, 100 kc/s to 120 megs. £12.50. Carr. £1.

BC211 with correct charts in fine condition £15 ea. Carr. £1.

PANAX Pulse generator G100H. Mint. £40. Carr. £1.50.

BRAND NEW INSTRUMENTS HOUSING. Size 8 x 6 x 7" deep. Comprising of anodised aluminium front and rear linked frame with recessed light blue front and rear panels. Detachable dark grey vinyl covered aluminium covers. Price £3.87 ea. P. & P. 25p.

FIBRE GLASS PRINTED CIRCUIT BOARD. Brand new. Single side 1p per sq. in. Double sided 1p per sq. in. Cut to size (Max. 2 1/2" x 15"). Postage 5p per order.

BERCO miniature variac type 31C. 0-250V 1 amp. 2 5/16th" depth, 3 diameter. Complete with dial and pointer. As new £3. P. & P. 37p.

SEQUENTIAL TIMERS 240V synchronous motor 1/2 rpm. 12 cam operated 2 pole micro switches. Individually adjustable from 0° to 180°. £6 ea. Standard 240V MOTORS with reduction gearbox 14 lbs. per sq. inch. £3 ea. Modern replacement for VCR 138 tube. Flat face 3 in. £1.63. P. & P. 25p. Bases 17p.

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FIREBALL TURRETS. Brand new £1.50 ea. P. & P. 25p. Sub-miniature IFF's 465/470 kc/s. Size 1/2 x 1/2 x 1 1/2" high. Set of 3—£3p. Sub-min. Vitality bulbs 8V 1.2W 5 mm Clear L.E.S. 7p ea. 100 off 6p ea.

DUNFOSS—solenoid valves. 240V 50 c/s. Type EVJ 2. Brand new boxed £5; Second hand £3. P. & P. 6/-.

Precision THERMISTOR by YSI. 100 k. at 25°C. Range: 40°C to 150°C. Supplied with charts giving ohms for each degree over entire range. Brand new. £1.50 ea.

CLAUDE LYONS Main Stabilizer. Type TS-1L-5S0. Input 119-135 volts 47/65 cs. Output 127 +/- 0.25% 16 amps. £35. Carr. £2.

Panel mounting VARIAC 20 amp. 2 separate wipes (concentric shaft) £25. Carr. at cost.

ROBAND P.U. Type M99A. Stabilized 300 volts 2 amps. £22 inc. carriage.

E.H.T. Unit by Brandenburg model S.0530/10. £55.

KELVIN & HUGHES 4 channel recorder. £30 ea.

SMITHS twin channel recorder. Transistorised. £65. Various other single and twin track recorders from £20.

EVERSHED & VIGNOLES Recording paper. Brand new boxed. L618H 7" wide, 1 1/2" dia. 17p roll; 6" dia. £1 roll. JL900H 7" wide, 1 1/2" dia. 25p roll.

19in. Rack Mounting CABINETS 6ft. high 19in. deep. Side and rear doors. Fully tapped, complete with base and wheels. £12.50. Carriage at cost.

Double Bay complete with doors. Fine condition. £25. Carriage at cost.

TIME CALIBRATOR unit by Cawkell any or all time intervals from 0.5 microsecond to 1,000 microsecond. Internal calibration; gate generation £50. Carr. £1.50.

AUDIO/Vibrator Amplifiers 1 KW. £150 ea. Matching vibrators for above 3 1/2" x 2 1/2" dia. Weight approx. 1 ton. £100 ea. Smaller units available.

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4 DIGIT RESETTABLE COUNTERS. 1000 ohm. coil. Size 1 1/2 x 1 x 4 1/2 in. As new, by Solvco of Geneva. £2.50 ea. As above but 350 ohm. £3.50 ea.

METERS—Model 3705. 25-0-25 micro amp. Scaled. -100-0-100. 5 1/4" x 4". £3 ea.

SANGO 50 micro amp 4" round. Brand new boxed. £1.38. P. & P. 38p.

SANGO 50 micro amp rectangular meter. Size 2 1/2 x 3" with 4 separate scales, lever operated, 0/6 white, 0/60 blue, 0/600 red and set zero. £1.75. P. & P. 17p.

RECTANGULAR WESTON 5" mirror back. Scaled 0-750 1 ma basic 30/- ea.; 100 micro amp scaled 0-50 £2.50. P. & P. 17p.

SANGO 50 micro amp 3" round meters. Ex brand new radiation equip. £1 ea. P. & P. 17p.

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First come, first served

AMERICAN oscilloscope type TS34/AP. Size 7 x 6 1/2 x 15 1/2" deep with viewing hood. Tested, good working condition. Ideal general purpose scope. 117 volt mains therefore only £12.50. Carr. £1.

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SOLARTRON Stub. PU AS516 & AS517. Circuits supplied. Fantastic value at £2 and £4 each.

VERY SPECIAL OFFER. AVO Transistor Analysers in superb condition ONLY £30 each.

SUPERB BUYS. Furzehill V200A Valve millivolt meter 10mv to 1 kv. £10 ea. Furzehill Valve volt-meter 378B/2. 10 mv to 100 volts £7 ea.

MEGA Ohm Meters—check earths, bonding etc. Ridiculous at £5 ea.

SUNVIC DC Amplifier type DCA1. Thermo-couple etc. £9 ea.

Genuine MULLARD Transistors/Diodes. Tested and guaranteed. OC41, 42, 76, 77, 83; OA5, 10. All at 5p ea. OC23—10p ea.

COMPONENT PACK consisting of 2-2 pole 2 amp push on/off switches, 1 double; 1-small double pole push on/off; 200 resistors 1 and 1/2 watt many high values. Fine value at 50p per pack. P. & P. 17p.

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HIGH PERFORMANCE 11 TRANSISTOR THREE  
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**BARGAIN  
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This is a really top performance, top quality solid state receiver packed with **SONY** know-how and backed by the outstanding reliability for which **SONY** are renowned. Now this outstanding set is available from Laskys at over 27% below the manufacturers list price making it without doubt the **NUMBER ONE SCOOP** of 1971! Just look at these outstanding features. Covers MW, LW and FM (VHF), 11 transistor circuit for high sensitivity and stability. Powerful output to 5" P.M. Dynamic speaker with rich clear tone quality. AFC for drift free VHF reception. Push button wave change selectors and tone control. Choice of three power sources—9V battery, household mains or car battery with suitable adaptors. Dial light for use in the dark. External jacks for earphone, tape recording, external power input and car aerial. Ultra modern styling and superb finish with padded leatherette covered cabinet for superior sound damping with chrome trim, strong carrying handle.



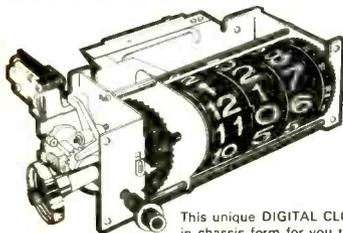
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## DIGITAL CLOCK MECHANISM



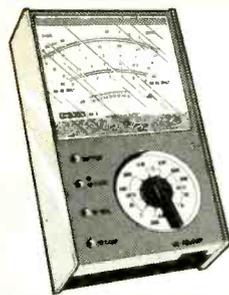
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This unique **DIGITAL CLOCK** is now available **EXCLUSIVELY** FROM **LASKY'S** in chassis form for you to mount in any housing that you choose. All settings are achieved by two dual-concentric controls at the front including: ON-OFF-AUTO and AUTO ALARM, "sleep" switch, 10 minute division "click" set alarm (up to 12 hour delay), time adjustment. Ultra simple mechanism and high quality manufacture guarantee reliable operation and long life.

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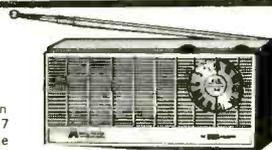
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- AC/V: 6-30-300-600 at 2.5K ohms/V
- DC Current 0-300mA 0-300mA
- Resistance: 0-10K ohms, 0-1M ohms
- Decibels: -10dB to 16dB
- Complete with test leads, battery and instructions

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The first pocket size receiver of its type allowing you to tune-in to the entire air communications band covered by 108-137 MHz in addition to full AM medium wave. Intermediate frequencies: AM 455 KHz; VHF 10.7 mc/s. Output power: 200mV 2 1/2 in. P.D. B ohm speaker. A built in ferrite rod aerial is provided for AM reception. The 10-406 is finished in blue with chrome trim, chrome telescopic antenna. Size 6 1/2" x 3 1/2" x 1 1/2 in. complete with batteries, magnetic earphone, instructions and circuit.

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**207 EDGWARE ROAD, LONDON, W.2.**  
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Ref. No.	VA (Watts)	Weight lb oz	Size cm.	Qty.		P.P. each
				1-24 Np	25-99 Np	
61	100	5 12	10.2 x 8.9 x 8.3	2.28	2.12	52
62	250	12 4	9.5 x 12.7 x 11.4	5.04	4.68	67
63	500	27 0	17.1 x 11.4 x 15.9	9.73	9.00	•
92	1000	40 0	17.8 x 17.1 x 21.6	17.93	16.58	•
128	2000	63 0	24.1 x 21.6 x 15.2	28.66	27.43	•
129	3000	84 0	21.6 x 21.6 x 20.3	46.38	42.89	•
190	6000	178 0	31.1 x 35.6 x 17.1	76.11	70.48	•



### AUTO SERIES (NOT ISOLATED)

Ref. No.	VA (Watts)	Weight lb oz	Size cm.	Qty.		P.P. each
				Auto Taps	1-24 Np	
113	20	11	7.3 x 4.3 x 4.4	0-115-210-240	0.74	0.69
64	75	1 14	7.0 x 6.4 x 6.0	0-115-210-240	1.43	1.32
4	150	3 0	8.9 x 6.4 x 7.6	0-115-200-220-240	1.74	1.61
66	300	6 0	10.2 x 10.2 x 9.5	•	3.38	3.12
67	500	12 8	14.0 x 10.2 x 11.4	•	5.02	4.64
84	1000	16 0	11.4 x 14.0 x 14.0	•	9.12	8.53
93	1500	28 9	13.5 x 14.9 x 16.5	•	13.12	12.22
95	2000	40 0	17.8 x 16.5 x 21.6	•	17.26	15.96
73	3000	45 8	17.4 x 18.1 x 21.3	•	23.47	21.73

### LOW VOLTAGE SERIES (ISOLATED)

Ref. No.	PRIMARY 200-250 VOLTS		Size cm.	12 AND/OR 24 VOLT RANGE		Qty. 1-24 Np	Qty. 25-99 Np	P.P. each
	Amps.	Weight lb oz		Secondary Windings	1-24 Np			
111	0.5	0.25	12	7.6 x 5.7 x 4.4	0-12V at 0.25A x 2	0.74	0.69	22
213	1.0	0.5	11	8.3 x 5.1 x 5.1	0-12V at 0.5A x 2	0.87	0.81	22
71	2	1	0	7.0 x 6.4 x 6.0	0-12V at 1A x 2	1.15	1.06	22
18	4	2	2	8.3 x 7.0 x 7.0	0-12V at 2A x 2	1.61	1.49	36
70	6	3	12	10.2 x 7.6 x 8.6	0-12V at 3A x 2	1.95	1.80	42
72	10	5	6	7.9 x 10.8 x 10.2	0-12V at 5A x 2	2.56	2.37	52
17	16	8	3	12.1 x 9.5 x 10.2	0-12V at 8A x 2	3.85	3.15	52
115	20	10	11	13.1 x 11.4 x 10.2	0-12V at 10A x 2	5.02	4.69	67
187	30	15	16	13.3 x 12.1 x 12.1	0-12V at 15A x 2	9.27	8.58	82

### LOW VOLTAGE MULTITAP SERIES (ISOLATED)

Ref. No.	PRIMARY 200-250 VOLTS		Size cm.	30 VOLT RANGE		Qty. 1-24 Np	Qty. 25-99 Np	P.P. each
	Amps.	Weight lb oz		Secondary Taps	1-24 Np			
112	0.5	1	4	8.3 x 3.7 x 4.9	0-12-15-24-30V	0.87	0.81	22
79	1.0	2	0	7.0 x 6.4 x 6.0	•	1.18	1.06	36
3	2.0	3	2	8.9 x 7.0 x 7.6	•	1.59	1.46	36
20	3.0	4	6	10.2 x 8.9 x 8.6	•	2.15	1.95	42
21	4.0	6	0	10.2 x 9.5 x 8.6	•	2.56	2.37	52
117	6.0	7	8	12.1 x 9.5 x 10.2	•	3.79	3.50	52
89	10.0	12	2	14.0 x 10.2 x 11.4	•	6.20	5.73	67

Ref. No.	Amps.	Weight lb oz	Size cm.	50 VOLT RANGE		Qty. 1-24 Np	Qty. 25-99 Np	P.P. each
				1-24 Np	25-99 Np			
102	0.5	1	11	7.0 x 7.0 x 5.7	0-19-25-33-40-50V	1.15	1.06	30
103	1.0	2	10	8.3 x 7.3 x 7.0	•	1.69	1.56	36
104	2.0	5	0	10.2 x 8.9 x 8.6	•	2.33	2.18	42
105	3.0	6	0	10.2 x 10.2 x 8.3	•	3.15	2.93	52
106	4.0	9	4	12.1 x 11.4 x 10.2	•	4.20	3.88	52
107	6.0	12	4	12.1 x 11.1 x 13.3	•	6.20	5.73	67
118	8.0	18	9	13.3 x 13.3 x 12.1	•	8.10	7.49	97
119	10.0	19	12	16.5 x 11.4 x 15.9	•	10.15	9.38	97

Ref. No.	Amps.	Weight lb oz	Size cm.	60 VOLT RANGE		Qty. 1-24 Np	Qty. 25-99 Np	P.P. each
				1-24 Np	25-99 Np			
124	0.5	2	4	8.3 x 9.5 x 6.7	0-24-30-40-48-60V	1.18	1.09	36
126	1.0	3	0	8.9 x 7.6 x 7.6	•	1.64	1.51	36
127	2.0	5	6	10.2 x 8.9 x 8.6	•	2.56	2.37	42
123	4.0	10	6	11.4 x 9.5 x 11.4	•	5.02	4.64	67
120	6.0	16	12	13.3 x 12.1 x 12.1	•	7.28	6.73	82
122	10.0	23	2	16.5 x 12.7 x 16.5	•	12.05	11.15	•

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45	1.5	1	9	7.0 x 6.0 x 6.0	1.18	1.07
8	4.0	3	11	10.2 x 7.0 x 6.3	1.77	1.63
86	6.0	5	12	10.2 x 8.9 x 8.3	2.80	2.46
148	8.0	6	4	8.9 x 10.2 x 10.2	3.04	2.81
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BP 02	= 7402 Quadruple 2-input Positive NOR Gates	23p	20p 15p
BP 03	= 7403 Quadruple 2-input Positive NAND Gates (with Open Collector Output)	23p	20p 15p
BP 04	= 7404 Hex Inverters	23p	20p 15p
BP 10	= 7410 Triple 3-input Positive NAND Gates	23p	20p 15p
BP 13	= 7413 Dual 4-input Schmitt Trigger	35p	32p 29p
BP 20	= 7420 Dual 4-input Positive NAND Gates	23p	20p 15p
BP 30	= 7430 8-input Positive NAND Gates	23p	20p 15p
BP 40	= 7440 Dual 4-input Positive NAND Buffers	23p	20p 15p
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Type No.	Function	Price
		1-24 25-99 100 up
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BP932	Expandable dual 4-input NAND buffer	25p 23p 20p
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BP935	Expandable Hex Inverter	25p 23p 20p
BP936	Hex Inverter	25p 23p 20p
BP944	Dual 4-input NAND expandable buffer without pull-up	25p 23p 20p
BP945	Master-slave JK or RS	35p 32p 29p
BP946	Quad, 2-input NAND	23p 20p 15p
BP948	Master-slave JK or RS	35p 32p 29p
BP951	Monostable	90p 80p 70p
BP962	Triple 3-input NAND	23p 20p 15p
BP9093	Dual Master-slave JK with separate clock	80p 75p 70p
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BP701C—SL701C	TO-5	8	OP Amp	63p 53p 45p
BP702C—SL702C	TO-5	8	OP Amp Direct O/P	63p 53p 45p
BP702—72702	D.I.L.	14	G.P.O.P. Amp (Wide Band)	53p 45p 40p
BP709—72709	D.I.L.	14	High Gain O/P Amp.	53p 45p 40p
BP709P— $\mu$ A709C	TO-5	8	High Gain O/P Amp.	53p 45p 40p
BP741—72741	D.I.L.	14	High Gain O/P Amp (Protected)	75p 60p 57p
$\mu$ A703C— $\mu$ A703C	TO-5	6	R.F.—I.F. Amp	63p 53p 45p
TAA263	TO-72	4	A.F. Amp	45p 35p 27p
TAA263S	TO-74	10	G.P. Amp	80p 75p 70p

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SOLARTRON OSCILLOSCOPE TYPE CD. 1015; D.C. to 21 Mc/s. YX amplifier, triggering facilities, external calibrator, display 3 1/2 in. tube. Mains 100-250V. Price £85 carr. £2.

**MARCONI SIGNAL GENERATOR TYPE TF-144G:** Freq. 85 Kc/s-25Mc/s in 8 ranges. Incremental:  $\pm 1\%$  at 1Mc/s. Output: continuously variable 1 microvolt to 1 volt. Output Impedance: 1 microvolt to 100 millivolts, 10 ohms 100mV - 1 volt - 52.5 ohms. Internal Modulation: 400c/s sinewave 75% depth. External Modulation: Direct or via internal amplifier. A.C. mains 200/250V, 40-100c/s. Consumption approx. 40 watts. Measurements 29 x 12 1/2 x 10 in. New condition. £45 each, carr. £1.50.

**TRIPLETT SIGNAL GENERATOR Model 1632:** Contains an R.F. Oscillator calibrated in 10 fundamental bands, covering a freq. of 100 Kc/s-120 Mc/s. Also a buffer amplifier and modulator stage, a metering system, crystal Oscillator stage, and a self-contained Heterodyne Detector. The wide frequency range covers broadcast, standard short-wave, T.V. and FM channels. Operates 115V a.c. 50/60 c/s. Output Meter 0-0.3 V. Controls: Ext. Mod.; Int. Mod.; CW; Het. Det.; Xtal.; AFO/put; RF Level; O/put Units; and O/put Multiplier. Slow and Fast motion dial. Price £12.50 very good second-hand cond. Carr. 75p.



**SOLARTRON PULSE GENERATOR GP1101.2:** Period—2 microsecs to 100 msec; Pulse Duration—1 microsec to 100 msec; Delay time—1 microsec to 10 msec. All continuously variable in 5 ranges with fine control. Accuracy  $\pm 10\%$ . Pulse Amplitude—0.5V-100V. Accuracy  $\pm 10\%$  continuously variable in 4 ranges with fine control. Double Pulses; Pre-Pulse; Triggering; Square Wave O/put; Squaring Amplifier. Input—100-250V, 50-60 c/s. New condition with Manual. Price: **£85** each + **£1.25** carr.

**USM-24C OSCILLOSCOPE:** 3 in. oscilloscope with 2c/s to 10Mc/s vertical response, and 8c/s to 800Kc/s horizontal response. Sensitivity 50 mv. rms/inch. Triggered sweep, built-in trigger pulses and markers. Mains input 115V, 50c/s. Complete with all leads, probes and circuit diagram. **£42.50** each, carr. **£2**.

**OS-46U OSCILLOSCOPE:** A general purpose oscilloscope suitable for measuring signals from 0-1000V d.c. to over 50,000 c.p.s. (Further details on request, S.A.E.) **£35** each, carr. **£1.50**.

**SIGNAL GENERATOR TS-510A/U:** (Hewlett Packard). A general-purpose signal generator designed to furnish signals with a very low spurious energy content, suitable for alignment of narrow-band amplitude modulated receivers. It may be amplitude modulated by internally generated sine waves or by externally applied sine waves or pulses. Freq. Range—10-420 Mc/s in 5 bands,  $\pm 0.5\%$  accuracy. Emission—AM, CW, Pulse. O/put Voltage—0.1V-0.5V, calibrated  $\pm 2$  db accuracy. Modulation—Internal 400, 1000 c/s (0-90%). Built-in Crystal calibrator (1, 5 Mc/s). Price: **£150** each, complete with transit case, manual and all leads; OR **£125** each, Sig. Gen. only. Carr. both types **£2**.

**SIGNAL GENERATOR TS-403B/U (or URM-61A):** (Hewlett Packard). A portable, self-contained, general-purpose test equipment designed for use with radio and radar receivers and for other applications requiring small amounts of RF power such as measuring standing-wave ratios, antenna and transmission line characteristics, conversion gain, etc. Both the output freq. and power are indicated on direct-reading dials. 115V, AC, 50 c/s. Freq.—1800-4000 Mc/s. CW, FM, Modulated Pulse—40-4000 pulses per sec. Pulse Width—0.5-10 microsecs. Timing—Undelayed or delayed from 3-300 microsecs from external or internal pulse. O/put—1 milliwatt max., 0 to -127 db variable. O/put Impedance—50  $\Omega$ . Price: **£120** each + **£2** carr.

**SIGNAL GENERATOR TYPE 902:** (P.R.D.). A portable, general-purpose, broadband, microwave signal generator designed for testing and maintenance of aircraft radio and radar receivers in the SHF band. The RF output level is regulated by a variable attenuator calibrated in dbm. The frequency dial is calibrated in Mc/s. Provision is made for external modulation. Power Supply—115V,  $\pm 10\%$  A.C., 50 c/s. Freq.—3650-7300 Mc/s. Internal Transmission—CW, Pulse, FM. External Transmission—Square Wave, Pulse. Power O/put—0.2 milliwatts. O/put Attenuator: -7 to -127 dbm. Load—50  $\Omega$ . Price: **£135** each + **£2** carr.

**TEST SET TS-147C:** Combined signal generator, frequency meter and power meter for 8500-9600 Mc/s. CW or FM signals of known freq. and power or measurement of same. Signal Generator: O/put -7 to -85 dbm. Transmission—FM, PM, CW. Sweep Rate—0.6 Mc/s per microsec. Deviation—0.40 Mc/s per sec. Phase Range—3-50 microsec. Pulse Repetition Rate—to 4000 pulses per sec. RF Trigger for Sawtooth Sweep—5-500 watts peak. 0.2-6 microsec. duration, 0.5 microsec pulse rise time. Video Trigger for Sawtooth Sweep—Positive polarity, 10-50V peak. 0.5-20 microsec duration at 10% max. amplitude, less than 0.5 microsec rise time between 90% and 10% max. amplitude points. Frequency Meter: Freq. 8470-9360 Mc/s. Accuracy— $\pm 2.5$  Mc/s per sec. absolute,  $\pm 1.0$  Mc/s per sec. for freq. increments of less than 60 Mc/s relative,  $\pm 1.0$  Mc/s per sec. at 9310 Mc/s per sec. calibration point. Accuracy measured at 25° C and 60 humidity. Power Meter: Input: +7 to +30 dbm. Output -7 to -85 dbm. Price: **£75** each + **£1** carr.

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**CRYSTAL TEST SET TYPE 193:** Used for checking crystals in freq. range 3000-10,000Kc/s. Mains 230V, 50c/s. Measures crystal current under oscillatory conditions and the equivalent parallel resistance. Crystal freq. can be tested in conjunction with a freq. meter. **£12.50** each, **£1** carr.

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Note: Minimum orders accepted 20 per type.

### TWO NEW OSCILLOSCOPES FROM RUSSIA



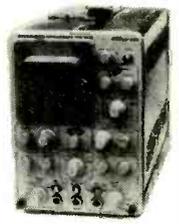
**CI-5 SINGLE BEAM OSCILLOSCOPE**

10 mc/s passband, triggered sweep from 1 µ sec. to 3 mill-sec. Free running time base from 20 c/s to 200 kc/s. Built-in time marker and amplitude calibrator, 3-in. cathode ray tube with telescopic viewing hood. £39.00

### CI-16 DOUBLE BEAM OSCILLOSCOPE

5 mc/s passband. Separate Y1 and Y2 amplifiers, rectangular 5 in. x 4 in. cathode ray tube. Calibrated triggered sweep from 0.2 µ sec. to 100 mill-sec. per cm. Free running time base 50 c/s to 1 mc/s. Built-in time base calibration and amplitude calibrator. £87.50

Full details on request. Full servicing facilities and spares available.



OUR NEW CATALOGUE 1970/1971 IS NOW READY. PLEASE SEND S.A.E. FOR YOUR FREE COPY.

0A2	0.38	5CP1	5.00	6BQ6GTB	
0A3	0.45	5D21	5.00		0.65
0A4G	1.15	5R4G	0.49	6BQ7A	0.85
0B2	0.38	5U4G	0.33	6BR7	0.85
0B3	0.60	5U4GB	0.42	6BR8	0.65
0C2	1.00	5V4G	0.42	6BR7	1.30
0C3	0.38	5W4GT	0.40	6BW6	0.85
0D3	0.35	5X8	0.50	6B7	0.70
1A3	0.30	5Y9GT	0.65	6BX6	0.25
1A5GT	0.32	5Z9	0.50	6B3XGT	
1B3GT	0.38	5Z4G	0.40		
1B4	4.00	6J30L2	0.75	6BZ6	0.75
1C5GT	0.35	6A8G	0.35	6C4	0.33
1CP31	8.00	6B4	0.35	6CB6	0.30
1G4GT	0.45	6AC7	0.25	6CD6GA	
1G6GT	0.40	6A7A	0.50		
1H5GT	0.42	6AG5	0.22	6CG7	0.50
1L4	0.20	6AG7	0.40	6CH6	0.55
1N5GT	0.48	6AH6	0.50	6CL6	0.50
1Q6GT	0.60	6AJ8	0.30	6CL7	0.35
1R4	0.25	6AK5	0.30	6CW4	0.45
1R5	0.35	6AK5W		6CV7	0.65
184	0.27	0.40	6D3	0.45	6N7GT
186	0.25	6AK6	0.57	6DC6	0.75
1T4	0.25	6AL3	0.43	6DK6	0.48
1T5GT	0.45	6AL5	0.20	6DQB	0.93
1U4	0.27	6AM3	0.32	6DB4	0.75
1U5	0.50	6AM6	0.33	6E5	0.50
1V2	0.45	6AN8	0.50	6EA8	0.50
1X2B	0.40	6AQ5	0.35	6EH7	0.38
2A3	0.40	6AQ6	0.55	6F7	0.35
2A21	2.25	6AR5	0.35	6F5	0.50
2C26A	0.90	6AR8	0.40	6FG6	0.30
2C39A	7.00	6AS5	0.35	6F11	0.38
2C40	3.50	6AS6	0.37	6F13	0.38
2C51	0.45	6A87G	0.50	6F14	0.45
2C63	4.50	6AT6	0.30	6F15	0.45
2C64	0.65	6AU6	0.25	6F17	0.50
2D21	0.25	6AV5GT		6F19	0.45
2E24	2.55	6B2	0.70	6F22	0.35
2K25	8.00	6AV6	0.30	6F23	0.80
3X2	0.30	6AW8A	0.55	6F24	0.75
3A4	0.35	6AX4GT		6F26	0.70
3B4	0.60	6B	0.45	6F28	0.60
3B28	2.15	6AX5GT		6GK6	0.60
3BP1	2.75	6J54	0.50	6H4	0.50
3D8	0.20	6B7	0.40	6J30	0.90
3D21A	2.00	6B8G	0.20	6K7	0.45
3Q4	0.40	6BA6	0.25	6K8G	0.35
3Q5GT	0.45	6BE6	0.30	6K8GT	0.50
384	0.35	6BF5	0.30	6K23	0.55
3V4	0.45	6BF6	0.50	6K25	0.70
4-400A		6BF6	0.50	6L6GT	0.45
4B32	10.00	6BH6	0.45	6L7	0.40
4HA5	0.48	6B16	0.45	6L18	0.45
4THA	0.45	6BK4B	1.20	6L20	0.40
5A8A	0.50	6BK7A	0.55	6ND7G	0.4c
5B/254M		6BL7GTA		6P1	0.60
	2.25	6P2	0.65	6P8	0.65
5B/255M	2.00	6BN6	0.40	6Q7	0.40
		6BQ5	0.25	6R7G	0.45

## FULLY GUARANTEED ZAERO BRAND FIRST QUALITY VALVES

2A7	0.30	30C15	0.80	328A	0.80	AC/HL/DD		E182CC	1.20	E180F	0.95	EL95	0.35	HBC90	0.30
2A7T	0.33	30C17	0.80	328A	1.00	AC/TH1	0.50	E186F	1.25	E181	0.45	EL95	1.15	HBC91	0.33
2A7U	0.30	30C18	0.75	328A	1.40	AX50	0.45	E188F	0.90	E182	0.35	EL803	1.00	HP93	0.35
2A7Y	0.55	30F5	0.8	328A	2.50	E280F	2.10	E189F	2.10	E183	0.65	EL821	0.55	HP94	0.30
2AV7	0.50	30FL1	0.70	328A	2.50	E281F	2.90	E190F	2.90	E184	0.55	EL822	0.90	HK90	0.35
2AX4GT	0.55	30FL2	0.93	328A	7.00	E282F	0.50	E191F	0.50	E185	0.55	EL823	0.75	HL23	0.40
2AX7	0.30	30FL3	0.80	328A	1.00	E283F	0.50	E192F	0.50	E186	0.40	EL824	0.75	HL23DD	
2AX7GT	0.70	30FL4	0.70	328A	1.50	E284F	0.55	E193F	0.50	E187	0.50	EL825	0.90	HL23DD	
2B14	0.55	30FL5	0.80	328A	1.50	E285F	0.55	E194F	0.50	E188	0.50	EL826	0.90	HL23DD	
2B14GT	0.35	30FL6	0.80	328A	1.50	E286F	0.55	E195F	0.50	E189	0.50	EL827	0.90	HL23DD	
2BA6	0.35	30P12	0.80	328A	3.50	E287F	0.55	E196F	0.50	E190	0.50	EL828	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E288F	0.55	E197F	0.50	E191	0.50	EL829	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E289F	0.55	E198F	0.50	E192	0.50	EL830	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E290F	0.55	E199F	0.50	E193	0.50	EL831	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E291F	0.55	E200F	0.50	E194	0.50	EL832	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E292F	0.55	E201F	0.50	E195	0.50	EL833	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E293F	0.55	E202F	0.50	E196	0.50	EL834	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E294F	0.55	E203F	0.50	E197	0.50	EL835	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E295F	0.55	E204F	0.50	E198	0.50	EL836	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E296F	0.55	E205F	0.50	E199	0.50	EL837	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E297F	0.55	E206F	0.50	E200	0.50	EL838	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E298F	0.55	E207F	0.50	E201	0.50	EL839	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E299F	0.55	E208F	0.50	E202	0.50	EL840	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E300F	0.55	E209F	0.50	E203	0.50	EL841	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E301F	0.55	E210F	0.50	E204	0.50	EL842	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E302F	0.55	E211F	0.50	E205	0.50	EL843	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E303F	0.55	E212F	0.50	E206	0.50	EL844	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E304F	0.55	E213F	0.50	E207	0.50	EL845	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E305F	0.55	E214F	0.50	E208	0.50	EL846	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E306F	0.55	E215F	0.50	E209	0.50	EL847	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E307F	0.55	E216F	0.50	E210	0.50	EL848	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E308F	0.55	E217F	0.50	E211	0.50	EL849	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E309F	0.55	E218F	0.50	E212	0.50	EL850	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E310F	0.55	E219F	0.50	E213	0.50	EL851	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E311F	0.55	E220F	0.50	E214	0.50	EL852	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E312F	0.55	E221F	0.50	E215	0.50	EL853	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E313F	0.55	E222F	0.50	E216	0.50	EL854	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E314F	0.55	E223F	0.50	E217	0.50	EL855	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E315F	0.55	E224F	0.50	E218	0.50	EL856	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E316F	0.55	E225F	0.50	E219	0.50	EL857	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E317F	0.55	E226F	0.50	E220	0.50	EL858	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E318F	0.55	E227F	0.50	E221	0.50	EL859	0.90	HL23DD	
2BA7	0.35	30P19	0.80	328A	17.00	E319F	0.55	E228F	0.50	E222	0.50	EL860	0.90	HL23DD	
2BA7															

# APPOINTMENTS VACANT

**DISPLAYED SITUATIONS VACANT AND WANTED:** £8 per single col. inch.

**LINE advertisements (run-on):** 45p per line (approx. 7 words), minimum two lines. Where an advertisement includes a box number (count as 2 words) there is an additional charge of 25p.

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**BOX NUMBERS:** Replies should be addressed to the Box number in the advertisement, c/o Wireless World, Dorset House, Stamford Street, London, S.E.1.  
No responsibility accepted for errors.

Advertisements accepted up to  
**THURSDAY, 12 p.m., 8th APRIL,**  
for the **MAY** issue, subject to  
space being available.

## Straight talking electronics engineers

Listen to us for a few well-paid months, then with computer expertise added to your thorough understanding of general electronics, you'll be a well qualified Service Engineer Instructor.

We're looking for that rare ability to make others see exactly what you're getting at. We want people who know their stuff inside out—who can pass on practical information that trainees would otherwise take years of experience to acquire.

It will be your responsibility to make sure that when your pupils leave the Training Centre as computer service engineers, they're (almost) as good at their jobs as you are now at yours!

Some travelling will be involved in the UK, and possibly overseas, and during this time a salary premium is paid in addition to all normal expenses.

Most of you will be based at Letchworth in the pleasant Hertfordshire countryside, and only an hour's drive from London. Relocation expenses will be considered.

Please write, quoting ref WW666C to A. E. Turner, International Computers Limited, 85/91 Upper Richmond Road, Putney, London, SW15.

**International Computers**

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## RADIO OPERATORS

There will be a number of vacancies in the Composite Signals Organisation for experienced Radio Operators in 1971 and subsequent years.

Specialist training courses lasting approximately 8 months are held at intervals. Applications are now invited for the course starting in September 1971.

### Salary Scales

During training with free accommodation provided at the Training School:

Age 21	£848	per annum
" 22	£906	"
" 23	£943	"
" 24	£981	"
" 25 or over	£1,023	"

On successful completion of course:

Age 21	£1,073	per annum
" 22	£1,140	"
" 23	£1,207	"
" 24	£1,274	"
" 25 (highest agepoint)	£1,351	"

then by 6 annual increments to a maximum of £1,835 per annum.

Excellent conditions and good prospects of promotion. Opportunities for service abroad.

Applicants must be United Kingdom residents, normally under 35 years of age at start of training course, and must have at least 2 years operating experience or PMG qualifications. Preference given to those who also have GCE 'O' level or similar qualification. Exceptionally well qualified candidates aged from 36-40 may also be considered.

Interviews will be arranged throughout 1971.

Application forms and further particulars from:

**Recruitment Officer, Government Communications Headquarters, Oakley, Priors Road, CHELTENHAM, Glos., GL52 5AJ.** Tel: Cheltenham 21491 Ext 2270

92

## TEST TECHNICIAN

Required for Final Production Testing and Fault Finding of Digital Voltmeters, and analogue to digital converters.

Experience of similar work or of Digital Systems is essential. Qualifications to H.N.C. advantageous, although opportunities exist for completion of professional qualifications.

Full staff status including pension scheme and attractive salary.

Reply to: **Head of Test Section, Fenlow Electronics Ltd., Jessamy Road, WEYBRIDGE, Surrey.**  
Tel: Weybridge 48177.

1075

## Sea-going Radio Officers can now make sure of a shore job and good pay.

If you'd like a job ashore, at a United Kingdom Coast Station, the Post Office will start you off on £1,080—£1,360, depending on age, with annual rises up to £1,850. There are good prospects of promotion to higher posts, opportunities exist for overtime and you would receive additional remuneration for attendance during the late evenings, at night and on Saturday afternoons and Sundays.

You will need to be 21 or over, with a 1st Class Certificate of Competence in Radiotelegraphy issued by the Postmaster General or the Ministry of Posts and

Telecommunications, or a Radiocommunication Operator's General Certificate issued by the Ministry of Posts and Telecommunications, or an equivalent certificate issued by a Commonwealth administration or the Irish Republic.

Find out more by writing to:

The Inspector of Wireless  
Telegraphy,  
I.M.T.R.  
Wireless Telegraph Section (W.W.)  
Union House,  
St. Martins-le-Grand,  
London,  
EC1A 1AR.

Post Office  
Telecommunications

1084

# ASSISTANT ENGINEER GRADE II (BROADCASTING) BOTSWANA

- ★ Salary up to £2,387
- ★ Low taxation
- ★ Appointments grant £100 or £200 in certain circumstances
- ★ 25% gratuity on basic salary
- ★ Contract 24 – 36 months
- ★ Subsidised accommodation
- ★ Education Allowances

The Posts and Telecommunications Department requires an officer to undertake operational duties including the installation and maintenance of broadcasting equipment in transmitting stations and to assist with the training of junior engineering staff.

Candidates must possess the City and Guilds Intermediate Certificate (Telecommunications) or equivalent and have had five years relevant practical experience, (additional to any period of approved training) of technical broadcasting equipment including M.F. and H.F. transmitting equipment up to 10 KW.

During the postal strike please telephone the Crown Agents' Appointments Division (01-222 7730 Ext. 665) for full details and an application form, quoting reference M2K/690420/WF, or write to The Crown Agents, 'M' Division, 4 Millbank, London, S.W.1.

1102

# ENGINEERS

## SERVICE AREA PLANNING

The INDEPENDENT TELEVISION AUTHORITY is looking for engineers with a keen sense of responsibility and initiative to join a test team concerned with planning the expansion of our VHF colour television service.

It is expected that the men selected will be aged 23-35, qualified to H.N.C. level, have had experience in television, R.F. measurements or an associated field and feel they would like the opportunity of working throughout the United Kingdom enjoying an open air life. They should, in addition, be experienced drivers with a clean and current licence. The ability to climb aerial support structures would be an advantage. There are currently two vacant posts;

1. The more senior post for which candidates should have a good working knowledge of radio wave propagation and basic television principles. The job will also involve taking charge of a small team.

Salary—on a scale £2115-£2574.

2. For the junior post we would consider less experienced candidates who can show good potential.

Salary—on a scale £1623-£1965.

Although based at our Knightsbridge Headquarters the selected candidates will be expected to work in any part of the United Kingdom for periods generally not exceeding two to three weeks at a time.

If you are interested in either of these posts please write or telephone for an application form, quoting Ref. WW/1605, 1 or 2, to:



The Personnel Officer,  
INDEPENDENT TELEVISION AUTHORITY,  
70 Brompton Road, London, S.W.3.  
Tel. 01-584 7011. Ext. 482.

Closing date for completed application forms: 31st March, 1971.

1083

## CIRCUITRY DESIGNER

Experienced in design of analog and digital circuitry to join research team developing electronic music synthesisers. H.N.C. or graduate.

Write or 'phone: **ELECTRONIC MUSIC STUDIOS (LONDON) LTD**  
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1080

# Data Line Officer

BOAC's world wide computer system requires expert control. We need a Data Line Officer to complete a team responsible for the operation of the terminal system. The appointed candidate will be based at London (Heathrow) Airport working on a rotating shift basis. He will occasionally be expected to travel to operational locations in North America and Europe.

Candidates should have qualified in electronics or data communications to H.N.C. standard or City & Guilds full certificate, and those with GPO line transmission experience will be given preference. Candidates qualified to O.N.C. standard will be considered only if they have appropriate experience.

Salary, including London Weighting, is in the range £1789 to £2179 per annum plus shift allowance of £383 per annum. Additional benefits include facilities for holiday air travel.

Please contact:

Selection Services, BOAC, Comet House (S Block 1st Floor), London (Heathrow) Airport, Hounslow, Middlesex. (Adjacent to Hatton Cross), or phone 01-759 5511 extn 2637 for an application form.

1099



## TEST ENGINEERS

This rapidly expanding company requires Test Engineers for test and alignment of transistorised V.H.F. Communications equipment.

Applicants should be familiar with V.H.F. techniques and capable of fault diagnosis.

Salaries offered will be commensurate with knowledge and experience. 38½ hour week, 3 weeks paid holiday. Non-contributory pension scheme. Ample opportunities for promotion.

Write or telephone

R. Trimmer, Chief Inspector,

**DYMAR**

Dymar Electronics Ltd.,  
Colonial Way, Radlett Road,  
Watford, Herts.

Tel.: Watford 21297

## BUSINESS OPPORTUNITY

Earn a substantial extra income through a fascinating part-time business of your own that you could share with your wife and operate from your own home. This is an outstanding business opportunity with rewards exceeding £5000 per annum at the higher levels. We are looking for organisational and managerial ability.

Telephone for an appointment.

VISTA MARKETING MAIDENHEAD 28754  
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## East Suffolk Education Committee

### LOWESTOFT COLLEGE OF FURTHER EDUCATION

Principal: A. E. BODDY, B.Sc. (Econ.) F.R.G.S.

#### LECTURER II

required in connection with work associated with City and Guilds Technicians course biased towards RADIO and ELECTRONICS, and for the Crafts Studies course in ELECTRICAL and ELECTRONIC ENGINEERING City and Guilds 500 series.

Salary in accordance with the Burnham scale £1,947-£2,537, plus allowances for approved training and qualifications. Starting point within the scale determined by past teaching experience and/or appropriate industrial experience. Applicants with teaching experience, appropriate qualifications and industrial experience should apply to the Principal, Lowestoft College of Further Education, St. Peter's Street, Lowestoft, telephone Lowestoft 4177, for further particulars and application forms.

1073

## Cambridge Audio

a member of AIM ASSOCIATES CAMBRIDGE GROUP

Cambridge Audio Laboratories Ltd., has a reputation for making state-of-the-art advances in the field of high quality domestic and professional audio equipment. We are looking for an exceptional Senior Engineer to lead our F.M. tuner research and development program. The location is St. Ives, Huntingdonshire.

The company's policy is one of continuous research and development and thus, after completion of the existing F.M. tuner design project, the successful applicant will have ample opportunity to pursue more open-ended research in this field.

Candidates must already have had considerable experience in the relevant technology and should be sufficiently qualified to be capable of providing original and theoretically sound solutions to design problems.

We are prepared to pay well for the right man and the starting salary will be up to £2,500. In addition, relocation expenses will be paid where appropriate.

Travelling expenses will be paid to interviewees.

Please telephone or write without delay to Ian Quayle, Technical Manager, Cambridge Audio Laboratories Ltd., The River Mill, St. Ives, Huntingdon PE17 4EP telephone: St. Ives (04806) 2901

## BRITISH RELAY

### TELEVISION and RADIO DISTRIBUTION SYSTEMS

We are expanding our activities in the field of wired installations in hotels, both at home and overseas. For this,

### WE REQUIRE ENGINEERS

with the necessary specialist knowledge and experience, for duties which include:—

- SYSTEM PLANNING
- SCHEDULING and ESTIMATING
- INSTALLATION CONTROL
- COMMISSIONING

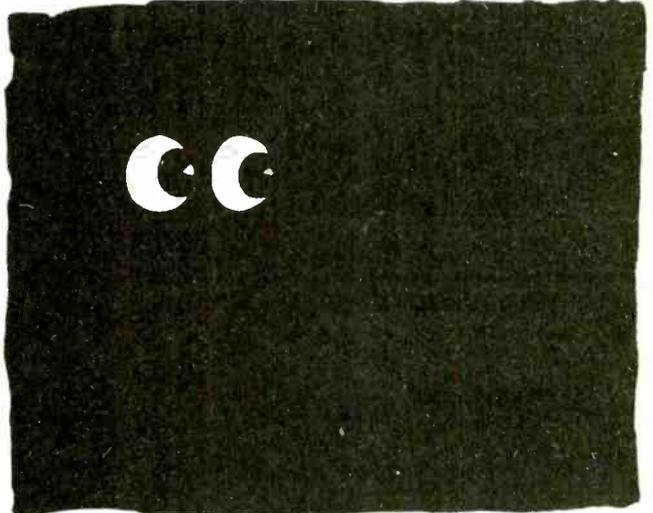
If you have experience which is relative to any aspects of this type of work, and would like information on staff vacancies, please apply to the address below.

*All enquiries will be treated in strict confidence.*

**THE GENERAL MANAGER,  
SPECIAL SERVICES DIVISION,  
British Relay House · 41 Streatham High Road ·  
London, S.W.16.**

**Tel: 01-677 9681**

## Light engineering/ electronics and in the dark about computers?



Join us now as a Computer Service Engineer, and after six months' paid specialist training, you will be responsible for ensuring that our computers are in peak condition.

We are Britain's leading computer manufacturer; we give men who want a rewarding career an excellent basic salary while we train them in every aspect of customer engineering in the computer industry. You'll learn to deal with operational problems, and to use the most intricate machinery.

HNC or C&G in electronics engineering, a Forces' training in electronics, or similar qualifications, are your passport to our opportunities.

How far you progress is up to you—the experience you get will stand you in good stead for your future career development. You'll gain knowledge of new methods and techniques on the most sophisticated equipment.

To add to your basic salary, you can get generous overtime and shift rates. There is a special allowance for working in central London. You will be operating in a computer environment on customers' premises in conditions well above the average for industry.

Age: 21/35.

Locations: Reading, Bracknell, Middlesex, Hertfordshire, Surrey, Central London, Manchester, Kidsgrove and Dublin.

Write giving brief details of your career, and quoting ref. WW668e to: A. E. Turner, International Computers Limited, 85/91 Upper Richmond Road, Putney, London SW15.

**International Computers**



# INSTRUMENT FOREMAN (DAYS)

required at  
**WYLFA NUCLEAR POWER STATION,  
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The Central Electricity Generating Board wishes to appoint to the above post an experienced electronics technician who has received a recognised formal training, including a City & Guilds Certificate or its equivalent, in a branch of instrument work.

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## University of Stirling

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#### SITUATIONS VACANT

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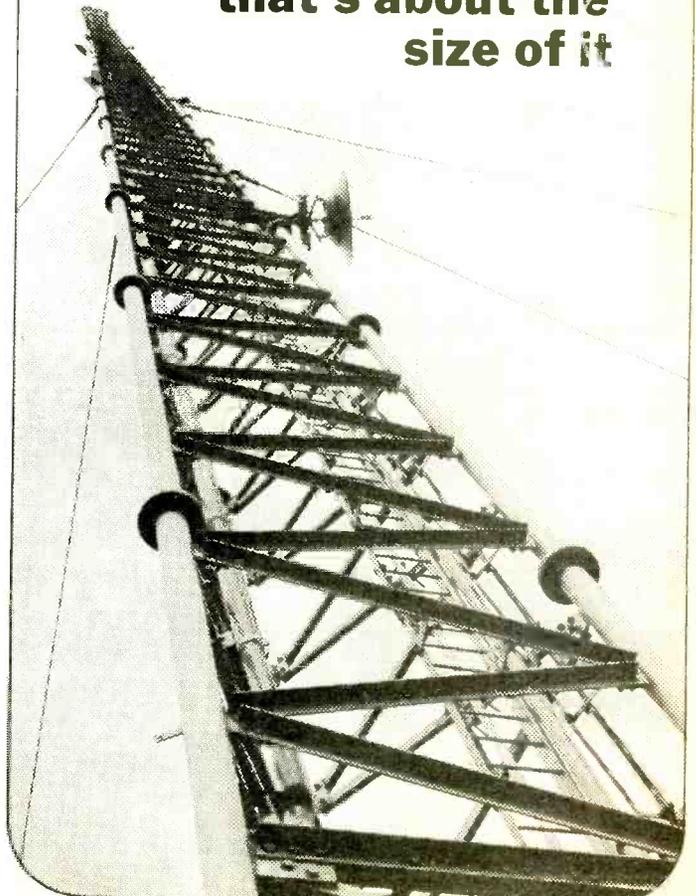
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ARMSTRONG 523	£40 4 6	£44 85			
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LEAK Teak	£40 9 0	£32 50			
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<h3>LAWSON TUBES</h3> <p>18 CHURCHDOWN RD. MALVERN, WORCS. Telephone: MALVERN 2100</p>	<table border="1"> <tr><th colspan="2">NEW TUBES</th></tr> <tr><td>14" £</td><td>4-50</td></tr> <tr><td>17" £</td><td>6-50</td></tr> <tr><td>19" £</td><td>7-50</td></tr> <tr><td>21" £</td><td>8-75</td></tr> <tr><td>23" £</td><td>9-75</td></tr> <tr><td>19" £</td><td>8-75</td></tr> <tr><td>23" £</td><td>11-95</td></tr> <tr><td>19" £</td><td>9-87</td></tr> <tr><td>23" £</td><td>13-50</td></tr> </table> <p>PANORAMA PANORAMA TWIN PANEL TWIN PANEL</p>		NEW TUBES		14" £	4-50	17" £	6-50	19" £	7-50	21" £	8-75	23" £	9-75	19" £	8-75	23" £	11-95	19" £	9-87	23" £	13-50	<table border="1"> <tr><th>REBUILT TUBES</th></tr> <tr><td>£4-25</td></tr> <tr><td>£4-87</td></tr> <tr><td>£5-25</td></tr> <tr><td>£6-87</td></tr> <tr><td>£7-25</td></tr> </table> <p><b>2 years Guarantee</b> both new and rebuilt <b>FULL TUBE FITTINGS</b> <b>INSTRUCTIONS SUPPLIED</b> CARR. INS. BY EXPRESS PASSENGER 14-19" 62p 21-23" 75p</p>	REBUILT TUBES	£4-25	£4-87	£5-25	£6-87	£7-25
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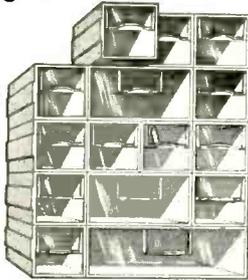
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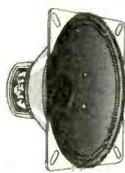
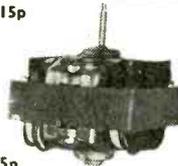
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500 mA MT 102 AT*	1.26 -224p	3A MT 105 AT	3.53 -374p	
1A MT 103 AT	1.88 -30p	4A MT 106 AT	4.55 -374p	
2A MT 104 AT	2.70 -30p	6A MT 107 AT	6.86 -50p	

Output Ref. No.	Price P.P.	60 Volts. All tapped at 0-24-30-40-48-60 V. Amps.	Output Ref. No.	Price P.P.
500 mA MT 124 AT*	1.33 -21p	2A MT 127 AT	2.83 -374p	
1A MT 126 AT	2.01 -30p	3A MT 125 AT	2.24 -374p	

Power output	Winding tapped at	Ref. No.	Price P. & P.
20 VA	0-115-210-240	MT 113 CT	£0.77 -104p
75 VA	"	MT 64 AT	£1.49 -274p
150 VA	0-115-200-220-240	MT 4 AT	£1.95 -274p
200 VA	"	MT 65 AT	£2.57 -274p
300 VA	"	MT 66 AT	£3.38 -35p
500 VA	"	MT 67 AT	£4.88 -45p

VA Ref. No.	Price P.P.	240 V. IN; 115 V. OUT; C.T.	VA Ref. No.	Price P.P.
60 MT 191 AT*	2.28 -30p	250 MT 194 AT*	5.80 -474p	
100 MT 192 AT*	2.48 -30p	350 MT 195 AT*	7.50 -59p	
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12-0-12	50 mA MT 210 C8**	£0.78 -74p
20-0-20	30 mA MT 211 C8**	£0.79 -74p
0-20 x 2	300 x 2 MT 214 CT**	£1.05 -15p
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0-15-27 x 2	500 mA x 2 MT 203 AT*	£1.98 -274p
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AT indicates open universal fixing with tags; CT is open U-clamp fixing with tags; C8 is open U-clamp fixing with P.C. spalls; \* with interwinding screen; † untapped 240V Primary; ‡ Primary tapped at 210-240V; other Primaries tapped at 200-220-240V.  
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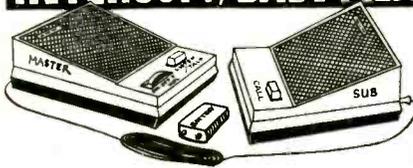
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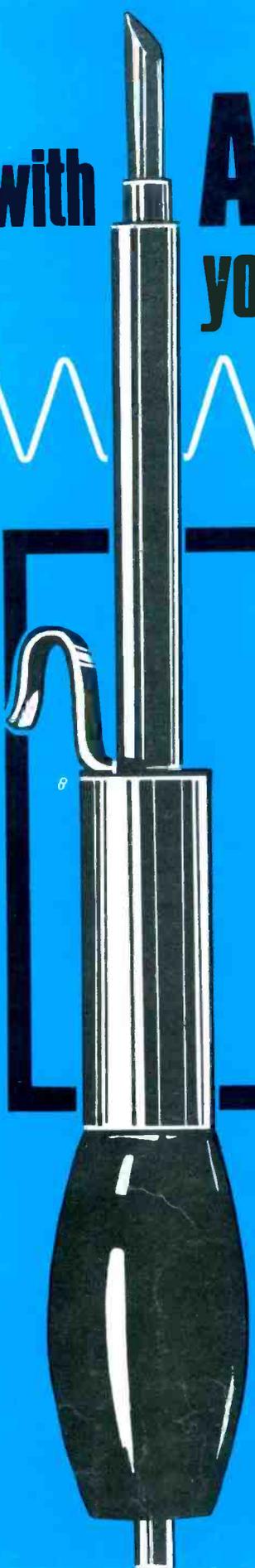
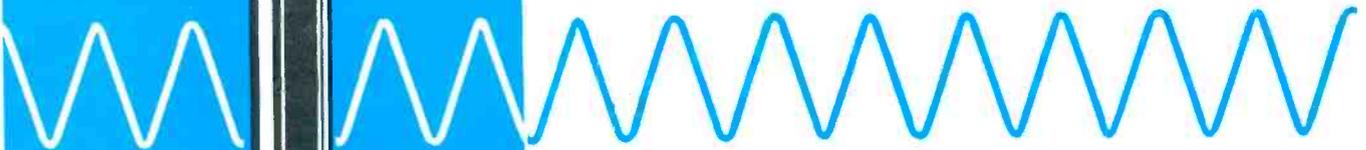
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