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The periodical, *Marconi Instrumentation*, which is mailed out free to all **mi** customers three times annually, contains technical articles about our latest instruments and up-to-the-minute information on their application to the solutions of measurement problems. It is written by engineers for engineers in English with summaries in four other languages.

And that's not all, by any means. **mi Contact** is a newspaper published six times a year to keep you

in touch with news and progress in the measurement business. Then there are our hardback publications, too. Already, there is a volume on TV Video Transmission Measurement written by the Head of BBC Measurement Systems Laboratory, and another book discusses the techniques and development of 'white noise' testing. Shortly we will be publishing a book on pulse code modulation, by a senior Post Office engineer.

There are technical data sheets, applications notes, catalogues, concise catalogues and product brochures, all aimed to help you measure.

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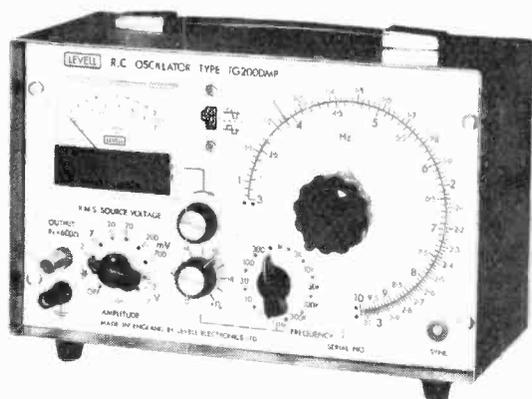
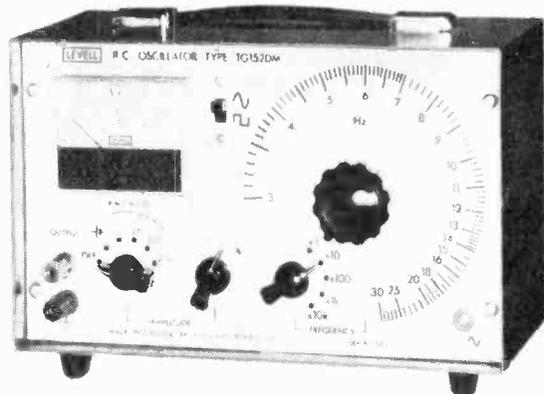
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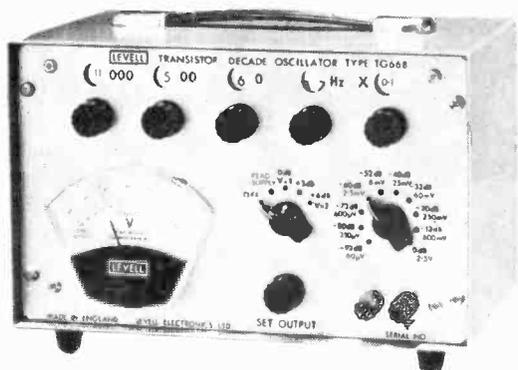
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r.f. test equipment

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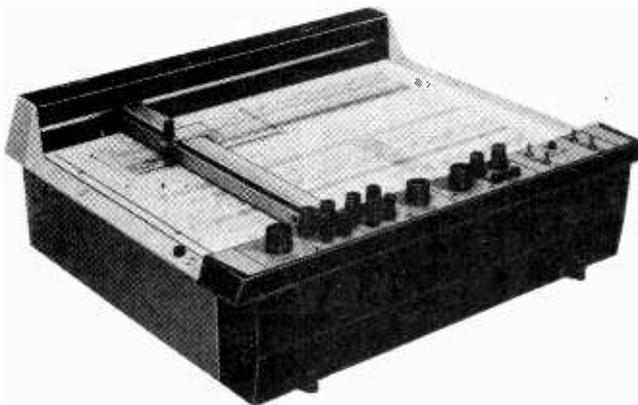
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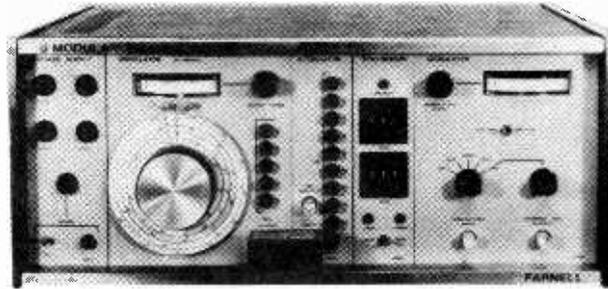
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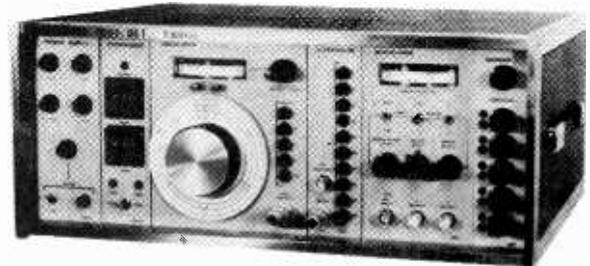


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Q1 1420 — SENIOR

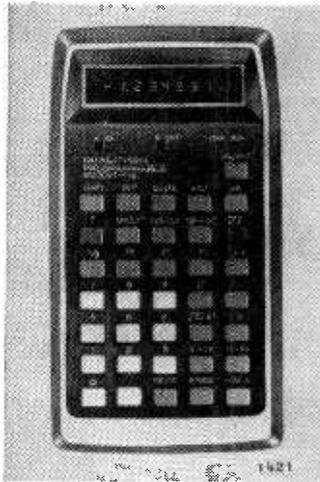


- Automatic selection of correct notation for result display (scientific or floating point)
- Dome keyboard for excellent response and preventing double entry input
- Algebraic mode operation
- Chain operations
- Change sign operation
- Three memories
- Display and memory exchangeable
- Trigonometric functions (sin, cos, tan)
- Inverse-trigonometric functions (\sin^{-1} , \cos^{-1} , \tan^{-1})
- Radian or degree selectable
- π constant
- Logarithms (ln, log)
- Anti-logarithms (e^x , 10^x)
- Combinatorial functions ($n!$, $\binom{n}{r}$, $(n)_r$)
- Normal distribution function ($Pr(x)$)
- Gamma function ($\Gamma(x)$)
- Group operations ($\Sigma \pm$, \bar{O} , \bar{X} , $\|x\|$)
- Group controls ($K \downarrow$, $K \uparrow$, $\Sigma \uparrow$, $\Sigma \downarrow$, CL_{GRP})
- Power function (y^x)
- Reciprocal ($1/x$)
- Square root (\sqrt{x})
- Square (x^2)
- Sum of squares (Σx^2)
- Summation (ΣX)
- Item count (n)
- Mean value (\bar{X})
- Mixed chain operations with parentheses approach (up to two levels)

- 14-digit LED display
- 10-digit mantissa with sign and 2-digit exponent with sign for data entry or results ($10^{99} \sim 10^{-99}$)

PRICE: £61.55 (Excluding VAT)

Q1 1421 — PROGRAMMABLE



- Reverse polish notation
- Display and Y-register exchangeable
- One accumulating memory (Memory store, Memory recall, $M + X$, $M - X$ and $M + X^2$)
- Trigonometric functions (sin, cos, tan)
- Inverse-trigonometric functions (\sin^{-1} , \cos^{-1} , \tan^{-1})
- Radians and degrees exchangeable
- π constant
- Logarithms (ln, log)
- Anti-logarithms (e^x)
- Power function (y^x)
- Reciprocal ($1/x$)
- Square root (\sqrt{x})
- Square (x^2)

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- Change sign operation

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Q1 1439—FINANCIAL



SPECIFICATION

- 9 digit green tube display
- 8 digits capacity for data entry or results ($10^7 \sim 10^3$)
- Fixed point (0, 1, 2, ... 7 places) or full floating point selectable
- Dome keyboard for excellent response and preventing double entry input

BASIC FUNCTION (+, -, x, %, /) AND MEMORY

- Algebraic mode operation
- Constant operations
- Repeat operations

- Chain operations
 - Percentage with automatic add on and discount
 - Change sign operation
 - Four memories
- SPECIAL FUNCTION**
- Financial operations
 - Cost, sell, margin operations
 - Trend line operations
 - Delta percentage operations
 - Item count (n)
 - Mean value (\bar{X})
 - Square root (\sqrt{x})
 - Summation (ΣX)
 - Sum of Squares (ΣX^2)

PRICE: £24.63 (Excluding VAT)

Q1 1419 — ADVANCED



SPECIFICATION

- 14 digit LED display
- 10-digit mantissa with sign and 2 digit exponent with sign for data entry or results ($10^{99} \sim 10^{99}$)
- Automatic selection of correct notation for result display (scientific or floating point)
- Dome keyboard for excellent response and preventing double entry input

BASIC FUNCTION (+, -, x, %) AND MEMORY

- Algebraic mode operation
- Constant operations
- Repeat operations
- Chain operations
- Change sign operation

- Display and Y register exchangeable
 - One accumulating memory
 - Display and memory exchangeable
- SPECIAL FUNCTION**
- Trigonometric functions (sin, cos, tan)
 - Inverse trigonometric functions (\sin^{-1} , \cos^{-1} , \tan^{-1})
 - Hyperbolic functions (sinh, cosh, tanh)
 - Inverse hyperbolic functions (\sinh^{-1} , \cosh^{-1} , \tanh^{-1})
 - Radian or degree selectable
 - π constant
 - Logarithms (ln, log)
 - Anti-logarithms (e^x , 10^x)
 - Power function (y^x)
 - Reciprocal ($1/x$)
 - Square root (\sqrt{x})

PRICE: £31.25 (Excluding VAT)

Q1 1444 — SLIDE RULE



SPECIFICATION

- 9-digit green tube display
- 8 digits capacity for data entry or results ($10^7 \sim 10^3$)
- Full floating point
- Floating negative sign
- Dome keyboard for excellent response and preventing double-entry input
- Automatic display blanking

BASIC FUNCTION (+, -, x, %, /) AND MEMORY

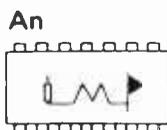
- Algebraic mode operation
- Constant operations
- Repeat operations
- Chain operations
- Percentage with automatic add-on and discount
- Change sign operation
- Display and Y-register exchangeable
- One accumulating memory

SPECIAL FUNCTION

- Reciprocal ($1/x$)
- Square root (\sqrt{x})
- Square (x^2)

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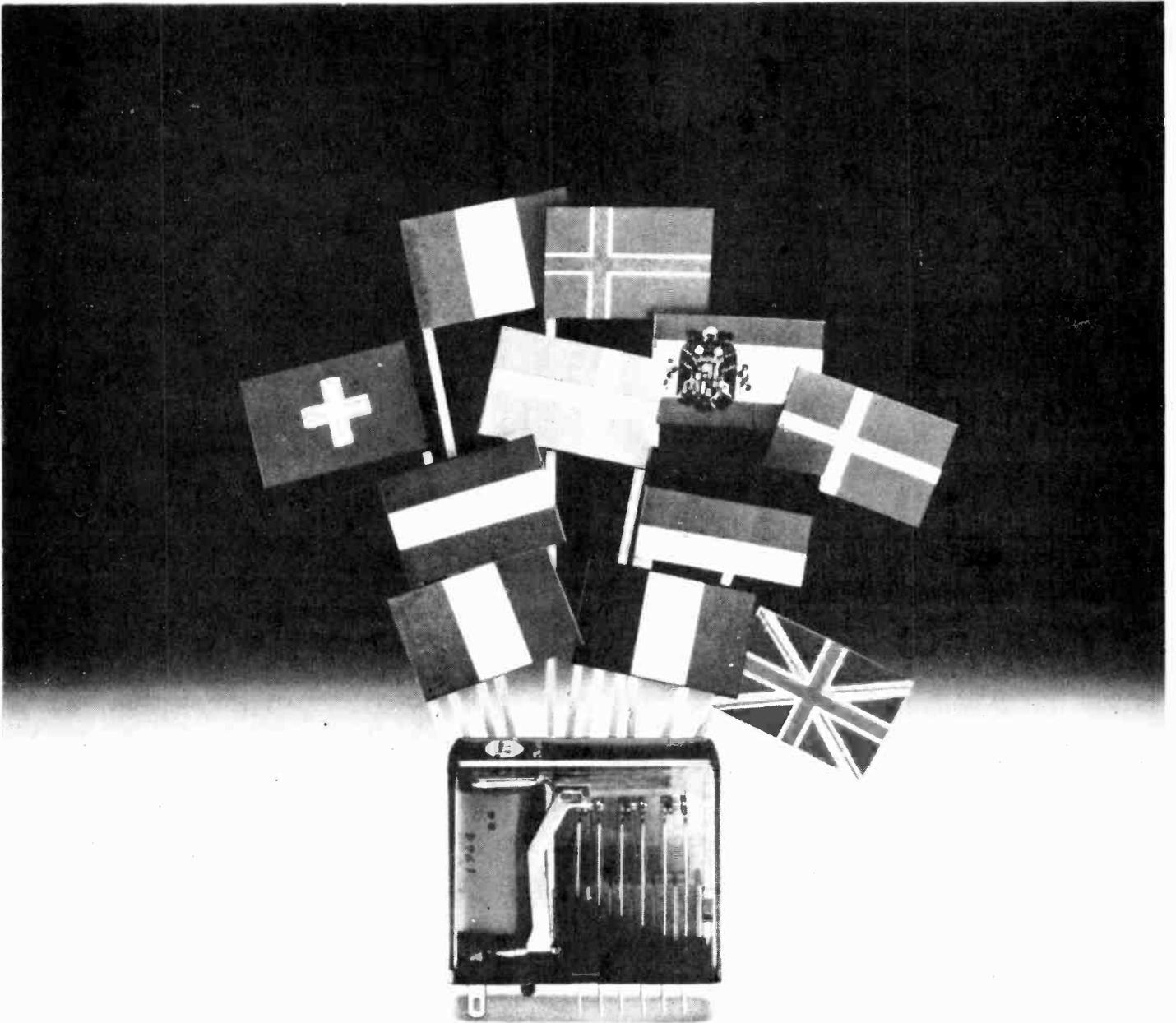
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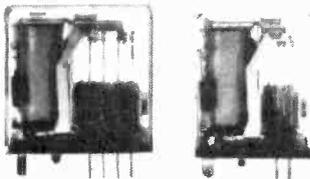
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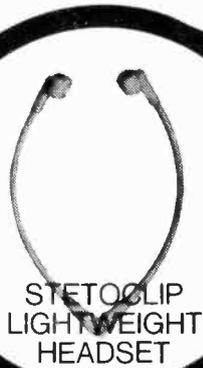
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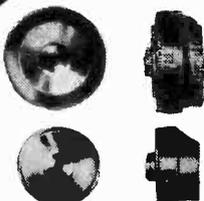
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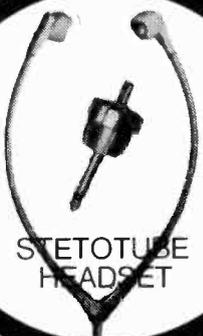
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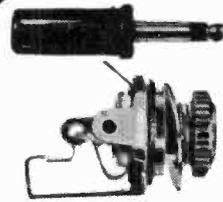
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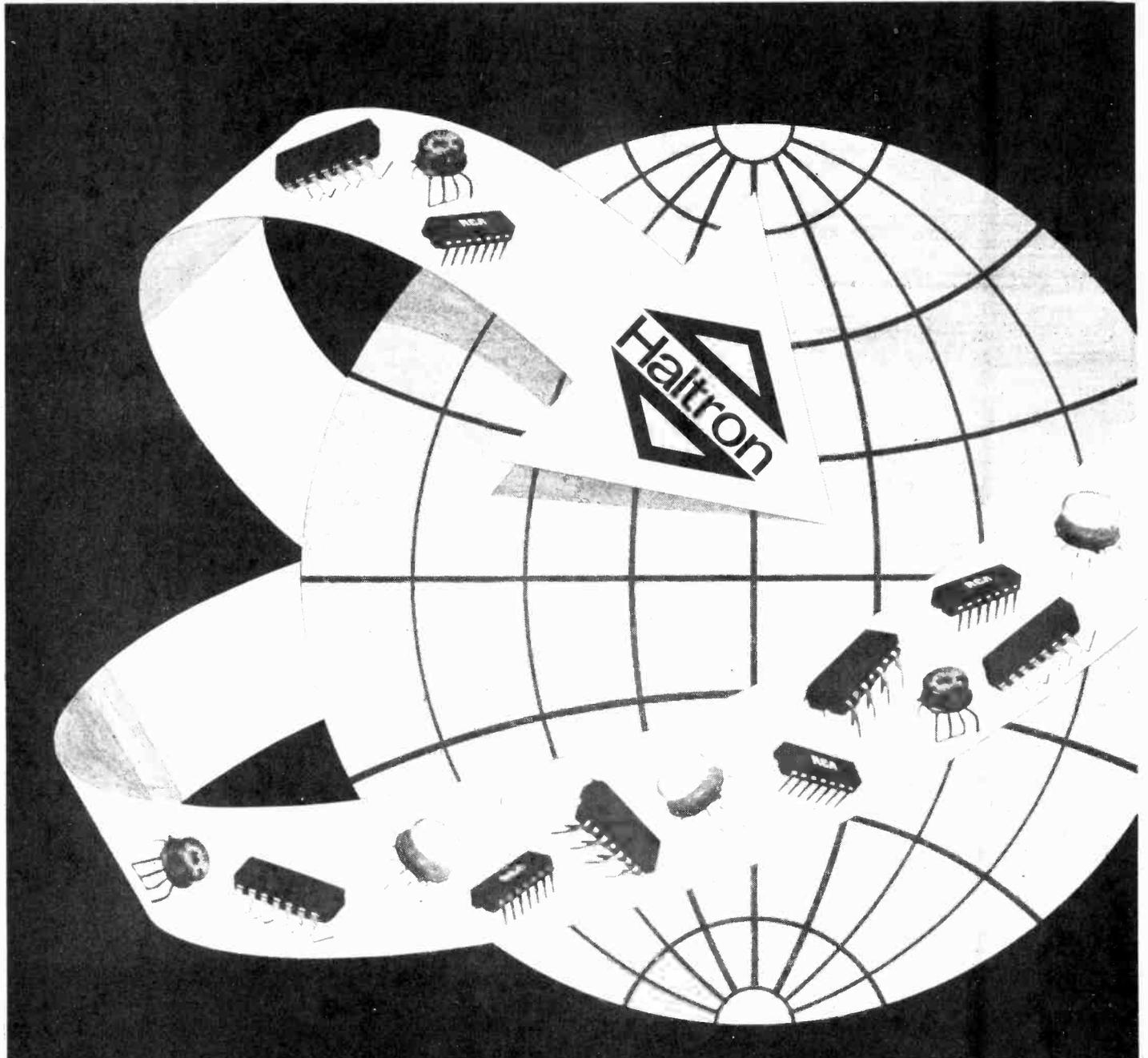
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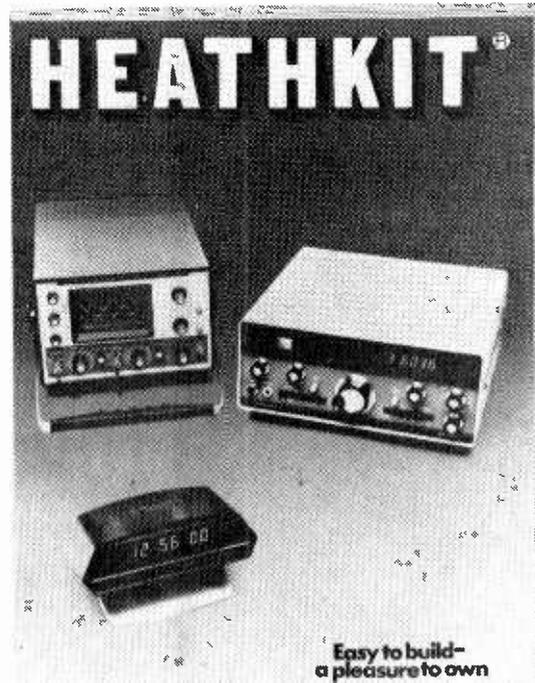
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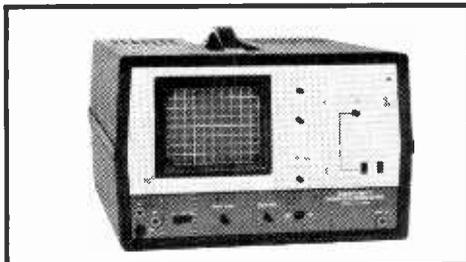
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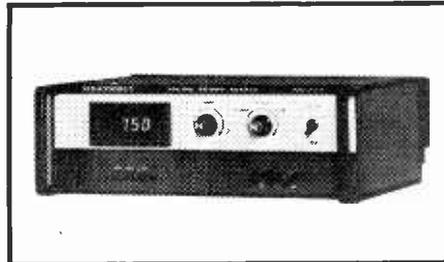
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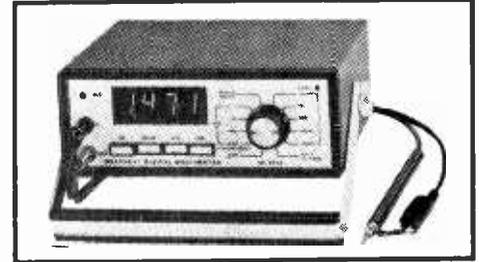
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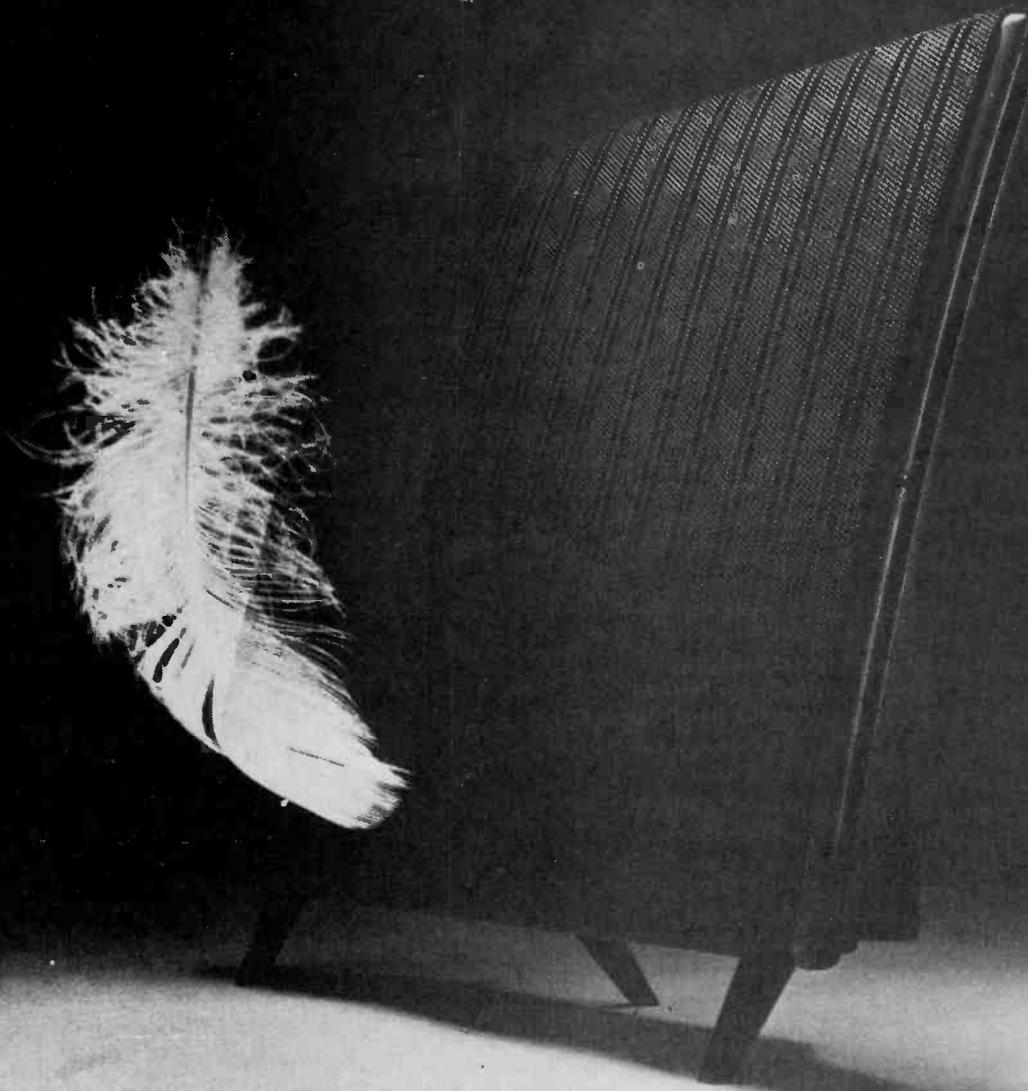
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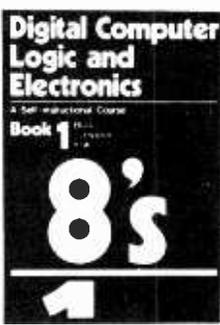
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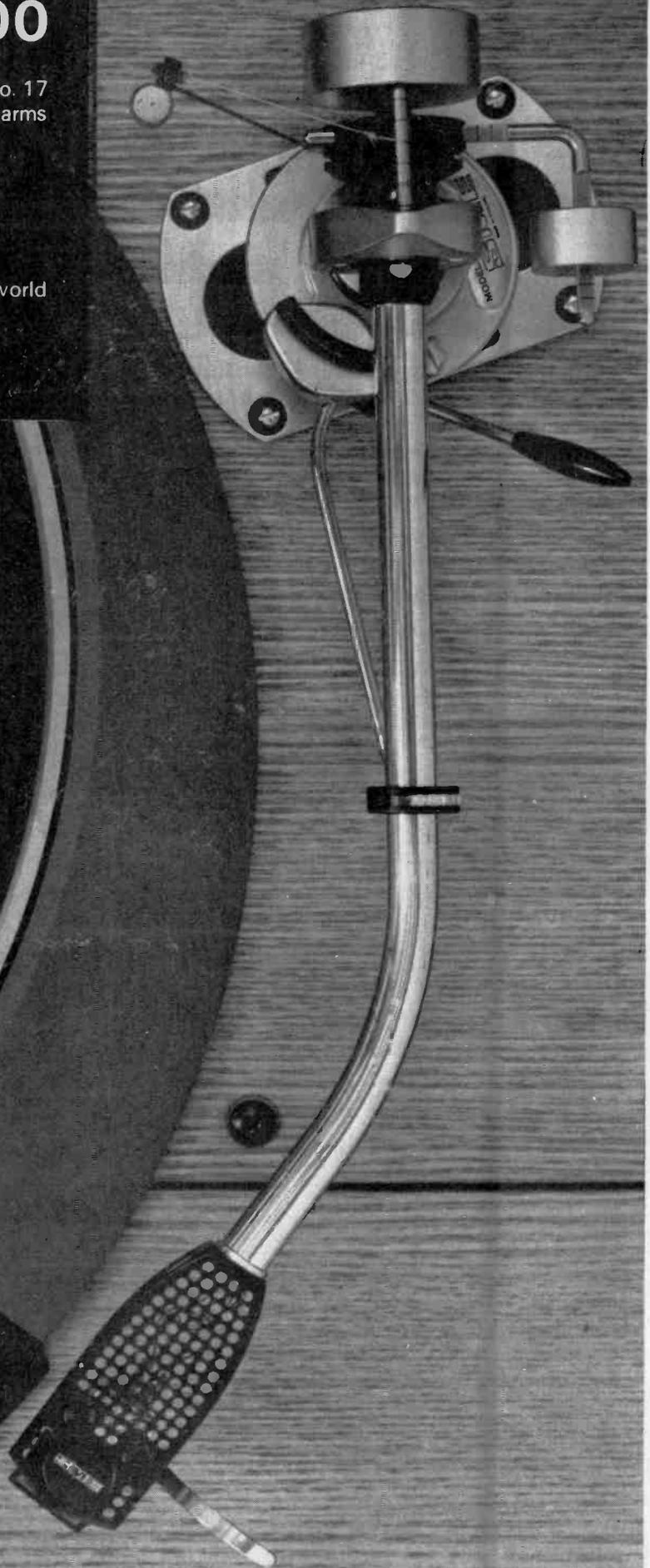
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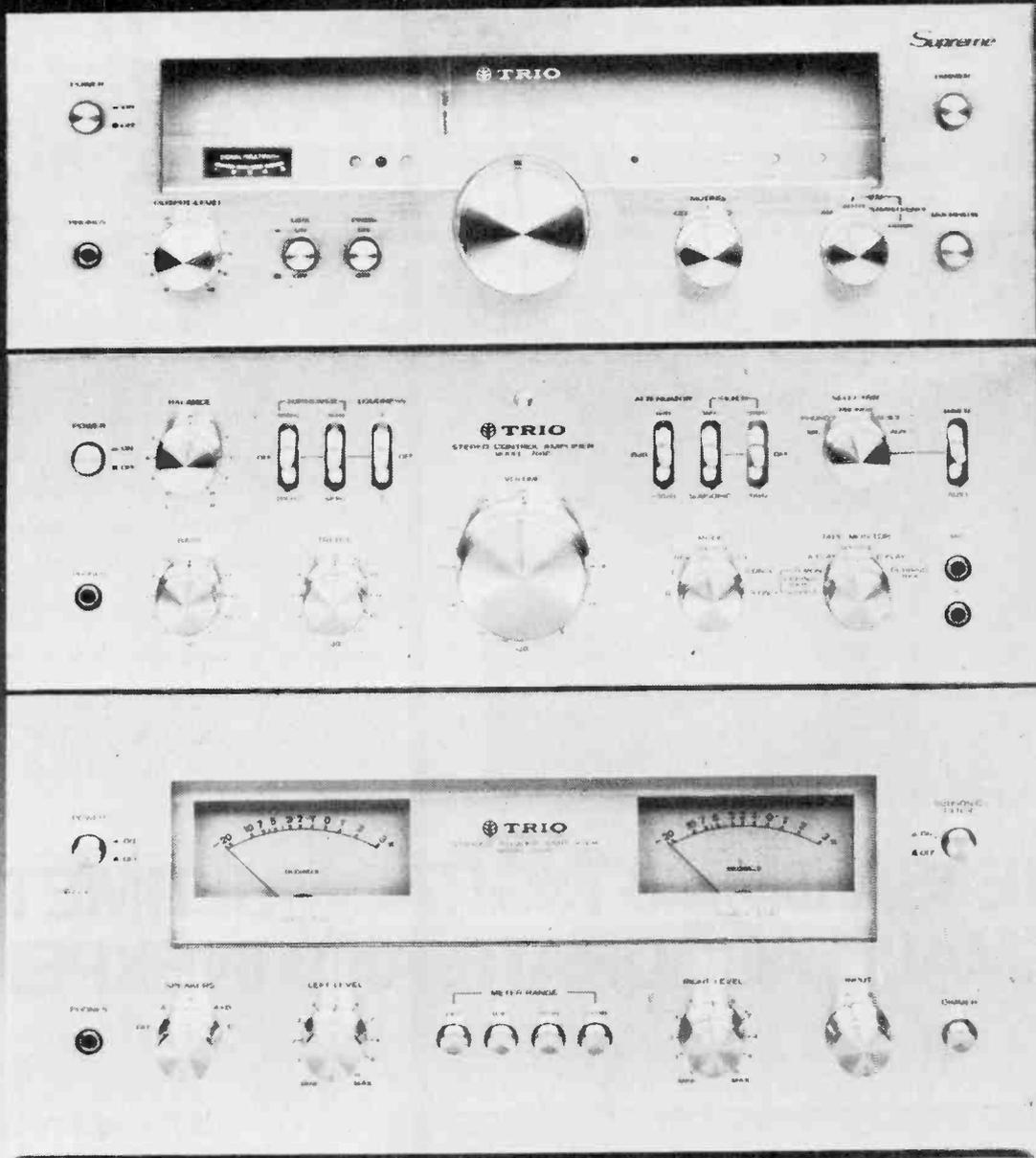
SME

The best pick-up arm in the world

Write to SME Limited
Steyning · Sussex · England

DENON DP-3000
DIRECT DRIVE TURNTABLE





£1,120 SOUNDS A LOT UNTIL YOU HEAR WHAT YOU GET FOR IT.

The trouble with equipment this advanced is there are so many features we haven't room for them here. But if you write to the address below we'll send you the details that will convince you £1,120 isn't a lot to ask.



The 700T stereo tuner £390.00. The 700C control amplifier £300.00. The 700M power amplifier £430.00. Prices shown are maximum selling prices excluding VAT. B. H. Morris & Co. (Radio) Ltd., Trio House, The Hyde, London NW9 6JP. Tel: 01-205 6441. Agents in Eire: Peat Wholesale Ltd., Dublin.

WW-061 FOR FURTHER DETAILS

More computer for your money from these upto date educational aids

Our family of low-cost computer educational aids is now larger than ever.

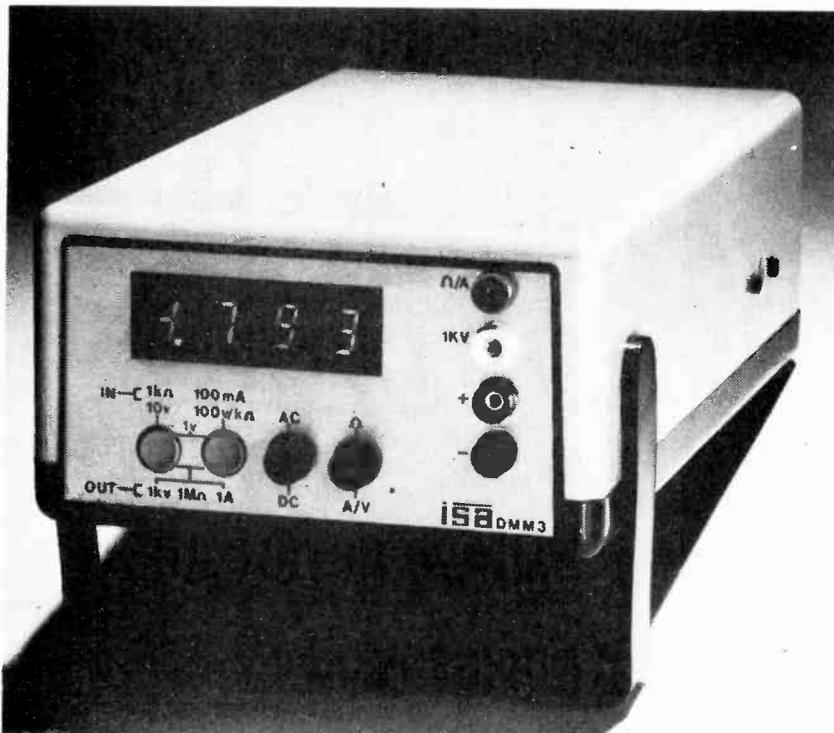
In addition to our popular Logic Tutors, we now have a fairly large selection of I.C. Patchboards, an Analog Tutor, a Digital Arithmetic Tutor and a unique Microprocessor Educational Kit.

You ought to get details.

LIMROSE ELECTRONICS LIMITED,
241-243 Manchester Road,
Northwich, Cheshire, CW9 7NE.
Tel. 0606 41696 and 41697

WW-058 FOR FURTHER DETAILS

THE ISA DMM3 DIGITAL MULTIMETER IT'S SMALL, ACCURATE, AND INEXPENSIVE



Now—a digital multimeter for the price of an analogue instrument.

At only 190mm x 105mm x 60mm it's easily portable, yet dual slope integration gives it an accuracy of $\pm 0.3\%$ at 2V DC to $\pm 1\%$ at 1000V AC. 15 ranges are covered—DC voltage at $\pm 2-2000$ with dual polarity and polarity indication, AC voltage 2-1000, AC DC amps $\cdot 2-2$, resistance 2K Ω /2M Ω . Read-out is presented on an 8mm LED display with a maximum reading of ± 1999 . Powered by 4 internal HP 2 batteries, with a life of up to 40 hours. Designed and manufactured entirely in the UK.

Price £47.50 plus VAT.

And currently, that's pretty good value.

Industrial Sub-Assemblies
isa

Trade enquiries invited.
Industrial Sub-Assemblies Ltd., 37 Telegraph Street,
Cottenham, Cambridge. Telephone (0954) 50590

The Greenwood guide to professional soldering.

Greenwood Electronics offer a range of highly advanced products specifically for professional soldering applications.

For more detailed information about the comprehensive Greenwood range, contact the address below.

1. The Iso-Tip. A safe, high-power iron which works anywhere without a mains lead. The breakthrough? Nickel Cadmium cells that are re-chargeable. (A charging stand is included for 240v or 115v A.C.) Each charge gives at least 60 soldering joints. Weight? Only 6 oz.

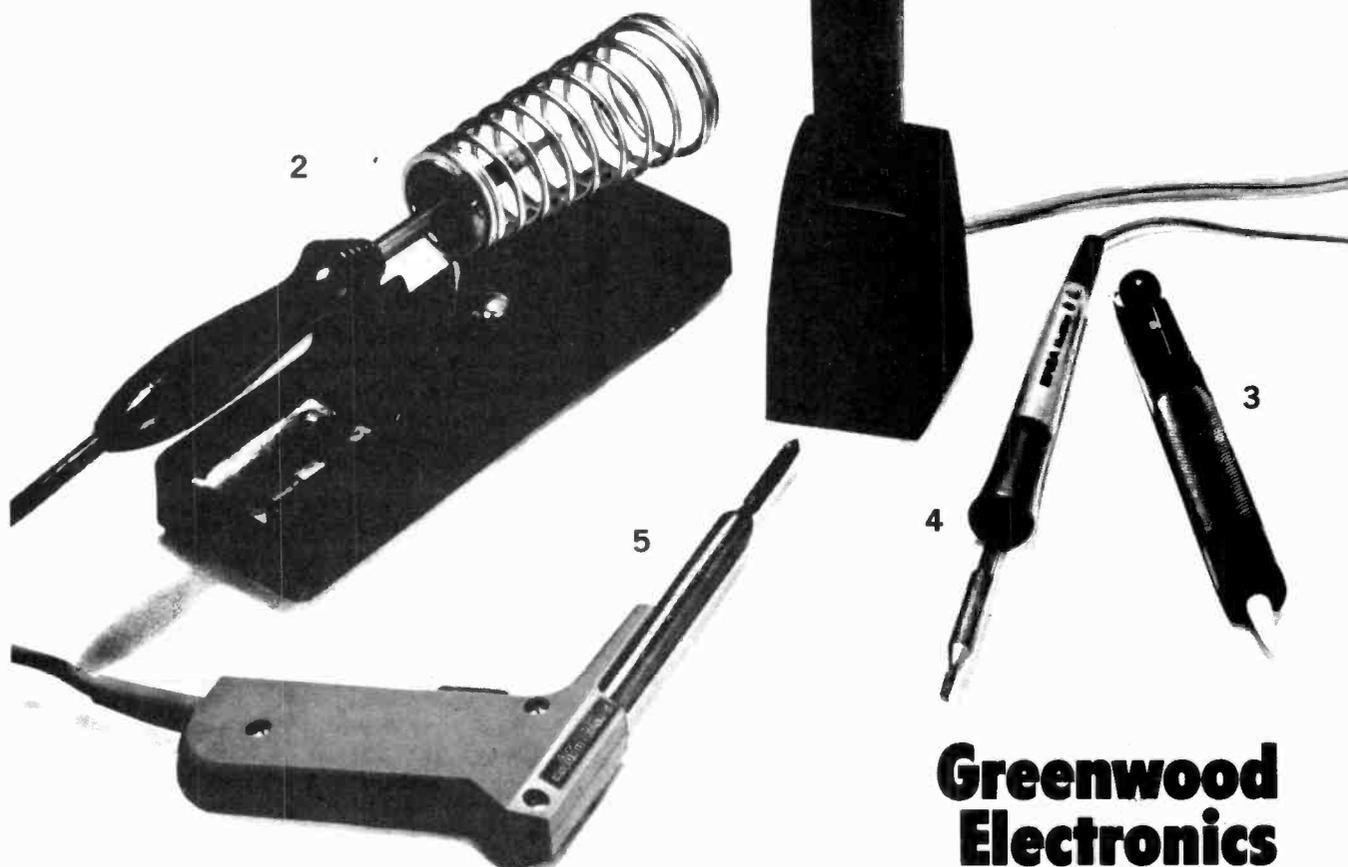
2. The Oryx 50. A temperature controlled mains soldering iron. (Temperature control within $\pm 2\%$). Adjustment (200° - 400°C) can be made whilst iron is operating, using the same tip. Light, compact, and easy to handle. A large 50W element loading gives rapid heating and high performance with constant tip temperature.

Also available: Oryx safety stand.

3. Oryx SR3A desoldering tool. Ideal where components are tightly grouped. Instantly removes unwanted solder from printed circuits etc. Accurate, reliable, speedy, and safe.

4. The Ersa Multitip. A top-quality iron that's ultra-light offering reliability so necessary to achieve constant production flow. A range of different shaped tips simply push onto the stem of the iron. It has the unique advantage that you can change the element in seconds.

5. The Ersa Sprint. Unique - it heats up to maximum temperature in only 10 seconds, and is the lightest gun on the UK market. Ideal for the service-man. With its light weight (only 7 oz.) and compact construction, it can be manoeuvred in even the most awkward areas.



Greenwood Electronics

Portman Road, Reading RG3 1NE. Tel: Reading (0734) 595844. Telex: 848659.



Now suitable for U.K., European and American voltages...

Minimod, the versatile British range of encapsulated power supplies first introduced in 1973, has now been extended to cover European and North American mains voltages (and is interchangeable with most American types). Normally available ex-stock, all units are fully stabilised with fold back current limiting — the 5V models have over voltage crowbar too!

STANDARD MODELS

Type Number	Output Voltage	Output Current Amps	Short Circuit Current mA (Typical)	% Regulation Line and Load (Typical)
PU01	5 ± 0.1	0.5	370	0.3
PU02	5 ± 0.1	1.0	770	0.5
PU03	15 - 0 - 15 ± 0.2	0.10	37	0.1
PU04	15 - 0 - 15 ± 0.2	0.20	84	0.1
PU05	12 - 0 - 12 ± 0.2	0.12	45	0.1
PU06	12 - 0 - 12 ± 0.2	0.24	120	0.2
PU11	18 - 0 - 18 ± 0.2	.15	50	0.1
PU10	15 ± 0.2	.10	37	0.1
PU12	12 ± 0.2	.10	45	0.1
PU13	18 ± 0.2	.065	23	0.1

Input voltage ranges 103 - 126V, 200 - 240V. 210 - 250V. Frequency 50 - 400 Hz all types.

Comprehensive specification given in brochure GT 29b which is available on request.

★ SPECIAL DESIGN SERVICE

Custom built units for applications requiring different specifications are produced as part of our standard service. Try us first.



Specialists in Electronic Transformers & Power Supplies.

GARDNERS

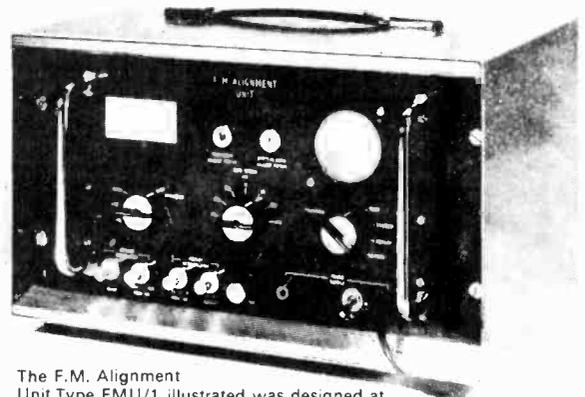
TRANSFORMERS LIMITED

Gardners Transformers Limited, Christchurch, Dorset, BH23 3PN. Tel. Christchurch 2284 (STD 0201 5 2284) Telex. 41276 GARDNERS XCH

WW—067 FOR FURTHER DETAILS

“Ampex and WHAT?...”

The JAMES SCOTT Alignment Units for D.R. and F.M. Multi-Channel Tape Recorders.



The F.M. Alignment Unit Type FMU/1 illustrated was designed at the Royal Radar Establishment, Malvern, to suit Ampex Recorders working on the IRIG intermediate band specification (using ES 100 Electronics) e.g. Model Numbers FR 1200, FR 1260, FR 1300, FR 1800L, FB 400, PR 500

If you have a sophisticated Ampex Recorder—Align it to the Manufacturers specification using our Alignment Units for D.R. & F.M. Systems.

Speedy and inexpensive

For Further information and Technical Literature Write or telephone.

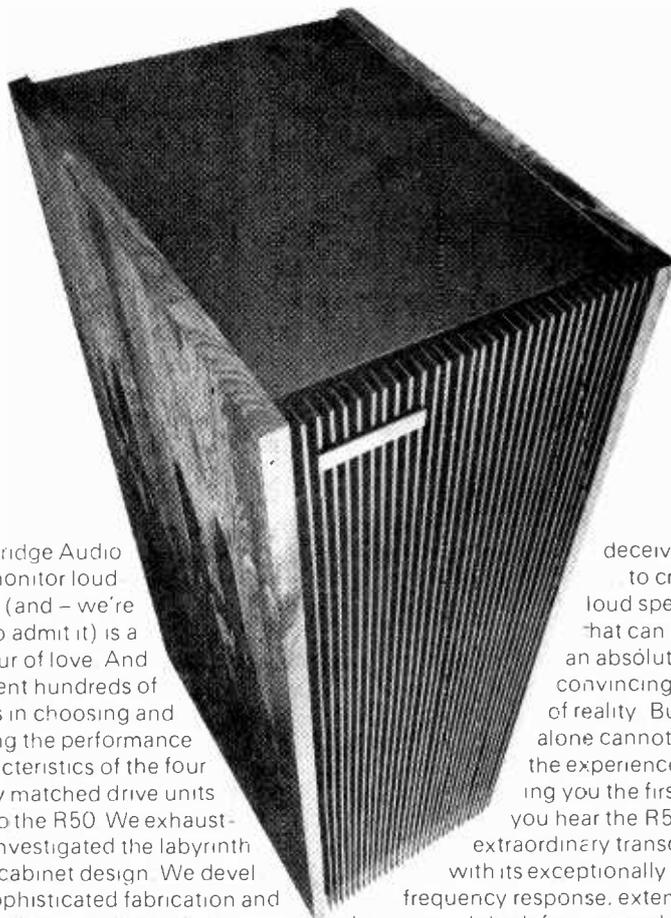


JAMES
SCOTT
(Electronic Engineering) Ltd

CARNTYNE INDUSTRIAL ESTATE
GLASGOW G32 6AB
Tel: 041-778 4206

WW—021 FOR FURTHER DETAILS

BEAUTIFUL ILLUSION OF THE THE ILLUSION OF REALITY



Our Cambridge Audio R50 monitor loud speaker (and – we're proud to admit it) is a true labour of love. And we've spent hundreds of hours in choosing and evaluating the performance characteristics of the four critically matched drive units that go into the R50. We exhaustively investigated the labyrinth paths of cabinet design. We developed sophisticated fabrication and testing techniques. In production we even go as far as to hand test and select each individual capacitor in the crossover network. In short, nothing is spared in our single minded effort to

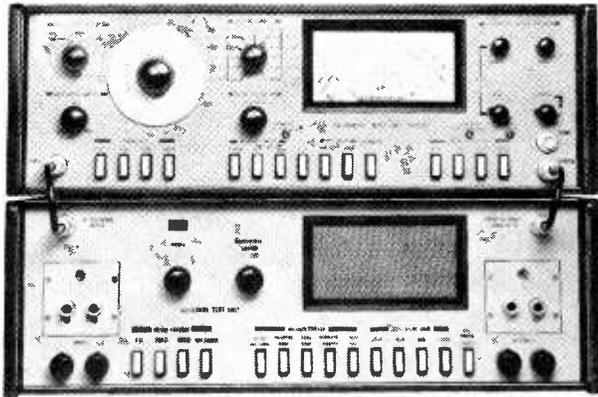
deceive you to create a loud speaker that can produce an absolutely convincing illusion of reality. But words alone cannot convey the experience awaiting you the first time you hear the R50. This extraordinary transducer with its exceptionally smooth frequency response, extended bass, superb high frequency dispersion and extremely low distortion has to be heard to be disbelieved. Only then will you begin to understand how close we have come to reality.



for people who listen to music
Cambridge Audio Limited
The River Mill
St. Ives
Huntingdon PE17 4EP
Telephone St. Ives 62901

WW-061 FOR FURTHER DETAILS

Audio Test Set



for amplifiers, mixers tape recorders

Checks . . . frequency response
signal/noise ratio
distortion
cross-talk
wow & flutter
drift
erasure
sensitivity
output power
gain
. . . in one compact unit.

Auxiliary Unit provides extra facilities for Studio testing.

Send for leaflet RTS2

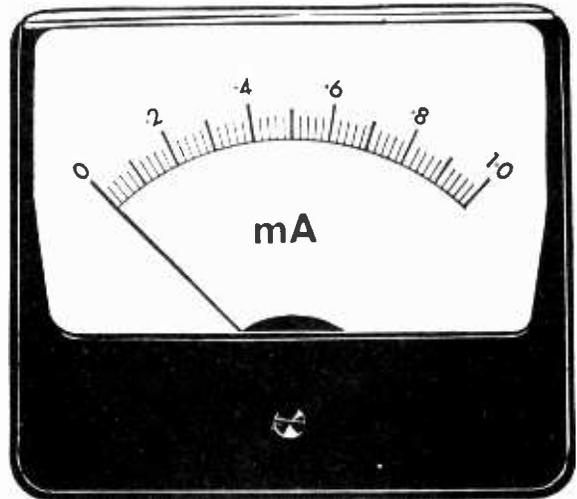
Ferrograph Company Limited Auriema House 442 Bath Road
Cippenham Slough Buckinghamshire SL1 6BB
Telephone: Burnham (062 86) 62511 Telex: 847297

FERROGRAPH

A member of the Wilmot Breedon group

WW-062 FOR FURTHER DETAILS

METER PROBLEMS?



137 Standard Ranges in a variety of sizes and stylings available for 10-14 days' delivery. Other ranges and special scales to order.

Full Information from:

HARRIS ELECTRONICS (London)

138 GRAYS INN ROAD, W.C.1 Phone: 01/837/7937

WW-039 FOR FURTHER DETAILS

Valradio

TRANSVERTORS

Valradio sinewave and square wave transvertors now incorporate SILICON transistors resulting in greater reliability and more stable performance at high ambient temperatures, including tropical climates.



TYPE D12/400S

A wide selection of types are available to drive practically any equipment within the power rating.

A random selection of types:

	Input	Output
C12/30S	12v DC	115/230v 30watts Sine wave
C24/60S	24v DC	115/230v 60watts Sine wave
D12/400S	12v DC	115/230v 400watts Sine wave
D12/500T	12v DC	115/230v 500watts Square wave
D24/150T	24v DC	115/230v 150watts Square wave
D12/250/24	12v DC	24v DC 8A

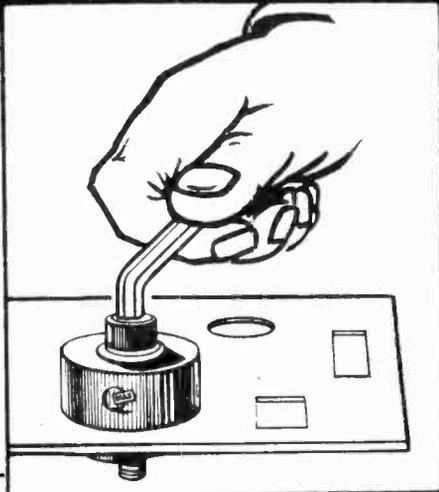
Please send for literature WW675 All prices +VAT

VALRADIO LIMITED
BROWELLS LANE, FELTHAM, MIDDLESEX, TW13 7EN
Tel: 01-890 4242/4837

WW-053 FOR FURTHER DETAILS



SHEET METAL PUNCHES FOR QUICK CLEAN HOLES



- Easiest and quickest way of punching holes in sheet metal (up to 1.625mm).
- Simple operation ● **100% British**
- Burr-free holes — no jagged edges
- **57 Metric and Linear sizes**

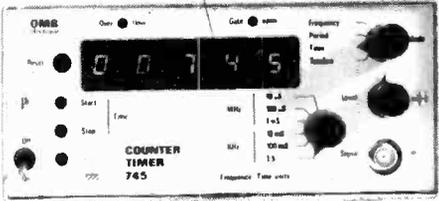
Used all over the world by: Government services — Atomic, Military, Naval, Air, G.P.O. and Ministry of Works; Radio, Motor and Industrial manufacturers, Plumbing and Sheet Metal Trades, Garages, etc.

Wholesale & Export enquiries to:

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WW-051 FOR FURTHER DETAILS

LOW-COST INSTRUMENTS

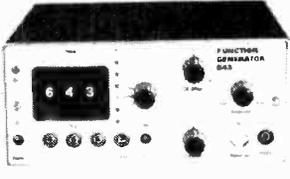


745 COUNTER TIMER
 Measures frequency, period, time and totalises
 32 MHz frequency range (DC coupled)
 5-digit .3" LED display
 6 Gate times/Time units, 10µs to 1 S in decades
 Sensitive, protected FET input

NEW!
£82
 + £1.50 p.p.
 + VAT



744 COUNTER TIMER
£74 + £1.50 p.&p. + VAT
 Measures frequency, period and time
 30MHz frequency range (DC coupled)
 5-digit, long-life incandescent display
 Sensitive, protected FET input



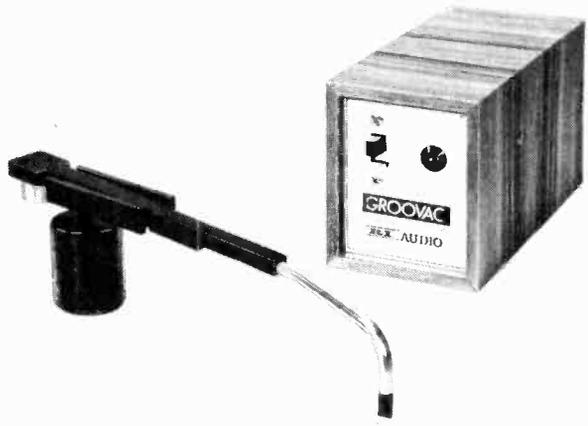
643 FUNCTION GENERATOR
£86 + £1.50 p.&p. + VAT
 Accurate, digital frequency setting
 .01Hz-1MHz
 Wide range external control of frequency
 Triangle, Squarewave and Low Distortion
 Sinewave outputs
 50Ω + simultaneous outputs
 DC offset

*Delivery is normally ex-stock—telephone for confirmation
 Prices correct at time of going to press, subject to change without notice*

OMB ELECTRONICS
 Riverside, Eynsford, Kent DA4 0AE
 Tel. Farningham (0322) 863567

WW-017 FOR FURTHER DETAILS

GROOVAC



vacuum record cleaner

Vacuum cleaning is the best way to remove dust, especially fine dust. Now with the Groovac, vacuum cleaning is available for extracting the particles from inside record grooves which are responsible for record and stylus wear — while your record is playing.

For full details please write to:—

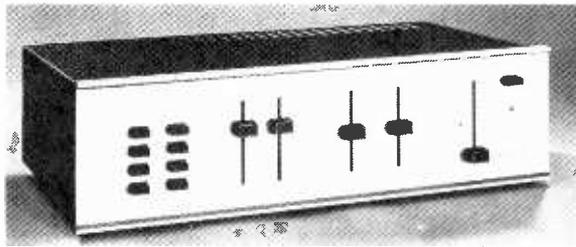


AUDIO Kernick Rd, Penryn
 Cornwall, England

WW-035 FOR FURTHER DETAILS

A NEW STANDARD FOR SOUND REPRODUCTION

HD250 High Definition Stereo Control Amplifier



Designed for disc and tuner input and two tape machines, with complete recording and reproducing facilities.

The HD250 amplifier establishes a new standard in amplifiers for sound reproduction in the home. Improvements have been made in respect of performance, engineering design and quality of construction. We believe that no other amplifier in the world can match the overall specification of the HD250. Look at extracts from the specification below.

Power output.

Rated: 50 watts average continuous power per channel, into any impedance from 4 to 8 ohms, both channels driven.

Maximum: 90 watts average power per channel into 5 ohms load.

Distortion.

Pre-amplifier: Virtually zero. (Cannot be identified or measured as it is below inherent circuit noise.)

Power amplifier: Less than 0.02% (typically 0.01% at 1kHz).

at rated output:
at 25w output:

Overload margin.

Disc input 40 dB min.

Hum and noise output

Disc: —83dBV Measured flat with noise bandwidth of 23kHz (ref. 5mV.)
—88dBV Measured with 'A' weighted characteristic (ref. 5mV.)

Line: —85dBV Measured flat (ref. 100mV.)
—88dBV 'A' weighted (ref. 100mV.)

Size: 17 inches \times 4 $\frac{3}{4}$ inches \times 11 inches deep overall.

Weight: 21 lb.

Write or phone for leaflet which describes the design philosophy and conception of the HD250 together with a complete specification.

RADFORD AUDIO LIMITED, BRISTOL, BS3 2HZ Telephone: 0272 662301

WW—016 FOR FURTHER DETAILS

Problem.

Where to obtain devices for push-pull audio power amplifiers which give good linearity and don't blow up on the slightest overload.

Solution.

M-OV audio beam tetrodes. A pair of KT66s will give up to 50W and a pair of KT88s will give up to 100W.

And M-OV audio triodes, too: a pair of DA42s gives up to 200W and a pair of DA 100s gives up to 300W.



LAP81

EEV and M-OV know how.

THE M-O VALVE CO LTD. Hammersmith, London, England W6 7PE.
Tel: 01-603 3431. Telex: 23435. Grams: Thermionic London. **S&C**

WW—008 FOR FURTHER DETAILS

Great Sound...



THE S3 PRESSURE UNIT has been designed to meet the growing demand for considerably increased power handling capacity without the sacrifice of either efficiency or frequency response. It features a powerful ceramic magnet and a strong but light diaphragm and voice coil assembly with many new features. It is a robust reliable unit of exceptional quality. The S3 is one of the units of the Vitavox Power Range.



VITAVOX
Limited
Westmoreland Road
London NW9 9RJ
Telephone: 01-204 4234

Please send me further information on your product range

Name
Company
Address

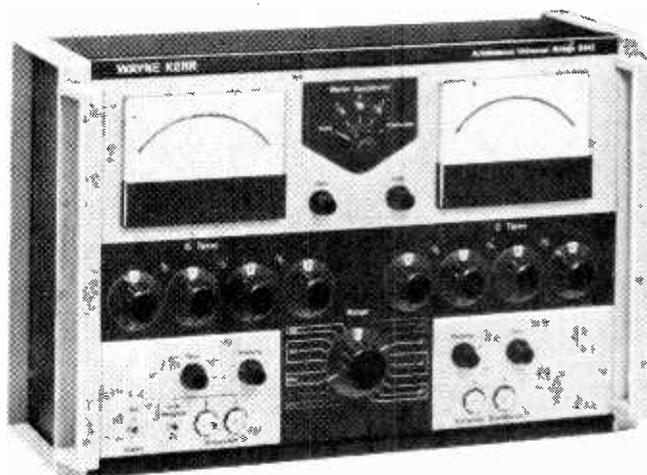
V/75/1

The world's most universal audio bridges

Wayne Kerr's B224 and B642



The B224 is a manually operated bridge, the resistive and reactive terms being independently set to a null indicated on the meter. A rechargeable battery is fitted in order to make the instrument portable.



The B642 balances itself automatically. The meters read real and quadrature terms and highly stable analogue outputs are provided which are directly proportional to capacitance and conductance above 10Ω impedance and also to inductance and resistance below 10Ω. One or two decades can be set to provide the first significant figures of the measurement, thereby increasing the meter sensitivity by 10 or 100 times. If a chart recorder is connected to the output of either term, drifts in component values to at least four significant figures can be observed.

For more information, telephone Bognor Regis on (02433) 25811 or write to the address below:

WAYNE KERR

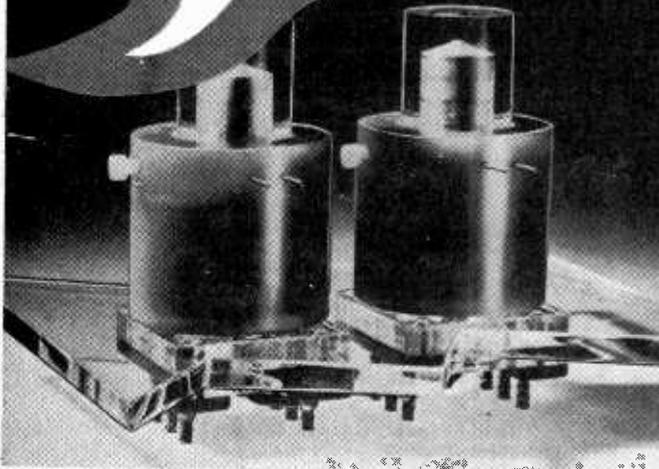
Durban Road, Bognor Regis, Sussex PO22 9R2
 Telex: 86120. Cables: Waynkerr Bognor
 A member of the Wilmot Breedon group

SPECIFICATION

Frequency	B224 (Manual balance)		B642 (Autobalance)	
		1592Hz (internal) 200Hz - 50kHz (external)		1592Hz (internal) 200Hz - 20kHz* (external)
Ranges for specified accuracy				
	0.1%	0.3%	0.1%	0.3%
C	100fF - 10μF	10μF - 10mF	1pF - 10μF	10μF - 10mF
L	1nΩ - 100mΩ	100mΩ - 1k	10nΩ - 100mΩ	100mΩ - 100Ω
H	1mH - 10kH	100nH - 1mH	1mH - 10kH	1μH - 1mH
R	10Ω - 1GΩ	1mΩ - 10Ω	10Ω - 100MΩ	10mΩ - 10Ω

NOTE: 0.1% accuracy relates to parallel component measurements above 10Ω impedance. 0.3% accuracy relates to series component measurements below 10Ω impedance
 *Manual operation only

*"He's asking
for a
reed relay
assembly
with a 30kV
isolated coil"*



People often bring their need to us. They know the Whiteley speciality. Being helpful! And the item that started life as a customer request, joins the Whiteley product list, ready to help other designers over a problem. You, perhaps? Consider a neat relay assembly — one or two dry reed switches with a rating of 25W, housed in a mounting tube, with either 'normally open' or 'change-over' contacts. Around them, a coil operating from 8, 12, 24 or 50V supply, 30kV isolated from the contacts. The whole unit mounting on a 0.25" insulating plate with a couple of 3 way tag strips. If you're interested, ask for a data sheet. But more, keep Whiteley in mind as the people who make useful things.

Surprising how often you'll find

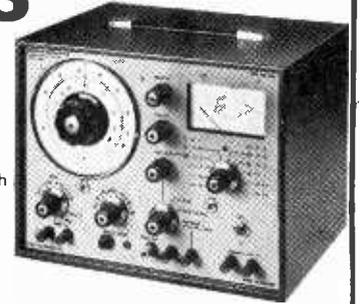
Whiteley make it.

Whiteley Electrical Radio Co. Ltd.

Mansfield, Notts NG18 5RW, England. Tel: 0623 24762.

WW—060 FOR FURTHER DETAILS

ROGERS AUDIO TEST EQUIPMENT



A comprehensive, versatile range of test equipment primarily designed for the measurement of high quality audio equipment but with additional applications in the electronics industry in general. The equipment is of particular interest to the professional audio engineer, recording studios, broadcasting authorities and educational establishments.

DM344A Distortion Factor Meter. Designed to make accurate and rapid measurements of total harmonic distortion generated within high quality audio amplifiers, recording and transmission equipment. **Selling Price: c/w Bench Case £175.00 + VAT.**

S324 Low Distortion Oscillator. Generates a pure sine wave and has been designed as a general purpose low distortion signal source. The primary application, used in conjunction with the DM344A, is the measurement of total harmonic distortion. **Selling Price: c/w Bench Case £80.00 + VAT.**

AM324 AF Millivoltmeter. Designed for voltage measurements in the audio and low RF ranges and principally for measuring low level signals in high impedance circuits. **Selling Price: c/w Bench Case £75.00 + VAT.**

PS1A. Regulated Mains Power Supply. **Selling Price: £18.50 + VAT.**



Model 'A' Noise Generator. A portable battery operated unit designed for carrying out listening tests on loudspeakers. 'Pink' or 'White' noise can be selected and output can be continuous or burst. Output is continuously variable. **Selling Price: £37.50 + VAT.**

Full Colour Literature describing the complete range may be had on request

ROGERS DEVELOPMENTS (Electronics) LIMITED
4/14 Barmeston Road, London SE6 3BN, England
Telephone: 01-697 8511 (3 lines)

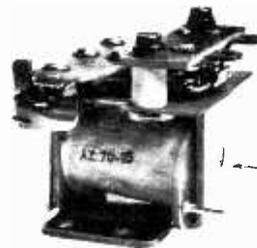
WW—030 FOR FURTHER DETAILS

Switching problems? Rely on Zettler.

Producing 30 basic types of relay and 15,000 variants with regard to contact stacks, terminals, energizing current and contact material, Zettler is among the largest manufacturers of electro-mechanical components.

Our product range comprises:

Low profile (flatform) ·
Timing · Miniature · Low
contact capacity · Herme-
tically sealed · Stepping ·
Mains switching · Latching
Contact stacks · Solenoids



**High-current Switching
Relays AZ 70... 72**

High contact rating, low energization requirement and fast switching times. 1 make, or 1 make and 1 break contact. AC switching capacity: 2500 VA. Max. voltage: 250 V AC. Max. current: 30 A. Coil voltages: 6 to 240 V DC.

We resolve your switching problems rapidly and expertly. Please contact us for further details.



ZETTLER Zettler
UK Division

Equitable House, Lyon Road
Harrow, Middx. HA1 2DU, Tel. (01) 8636329

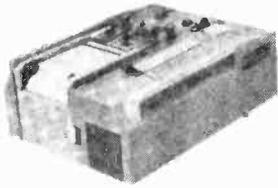
A member of the worldwide ZETTLER electrical engineering group, est 1877

Please look us up at the INTERNEPCON Exhibition, Brighton, 14-16 OCT., STAND No. 3314, Red Hall, Metropole Convention Centre.

WW—013 FOR FURTHER DETAILS

FAST RESPONSE STRIP CHART RECORDERS

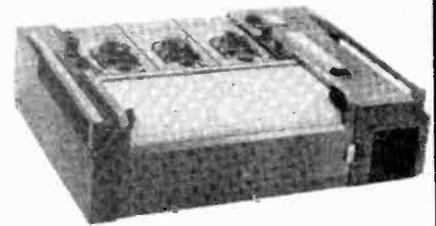
Made in USSR



Type H320-1
Single pen

Specification

Basic error 2.5%
Sensitivity 8mA F.S.D.
Response 0.2 sec.
Width of each channel 80mm
Chart speeds, selected by
push buttons 0.1—0.2—0.5—1—2.5—
—5—12.5—25mm/sec.
Chart drive 200-250v 50Hz



Type H320-3
Three-pen

Recording: Syphon pen directly attached to moving coil frame, curvilinear co-ordinates

Equipment: Marker pen, Timerpen, Paper footage indicator, 10 rolls of paper, connectors, etc.

Dimensions: H320-1: 285x384x16.5mm
H320-3: 475x384x16.5mm

PRICE: H320-1 £80.00
H320-3 £130.00
Exclusive of VAT

Available for immediate delivery

Z & I AERO SERVICES LTD.

44A WESTBOURNE GROVE, LONDON W2 5SF

Tel. 01-727 5641

Telex: 261306

WW—011 FOR FURTHER DETAILS

FREQUENCY COUNTERS

HIGH PERFORMANCE REASONABLY PRICED ELECTRONIC INSTRUMENTS



TYPE 901

CRYSTAL OVEN
TWO TONE BLUE CASE

£370 **520** MHz

Sensitivity 10mV. Stability 5 parts 10¹⁰

301M	32MHz 5 Digit £75	401	32MHz 6 Digit £118
501	32MHz 8 Digit £175	701	50MHz 8 Digit £185
701A	80MHz 8 Digit £195	801M	300MHz 8 Digit £285
901M	520MHz 8 Digit £370	Memory versions available if not suffixed M £25 extra	

Start/Stop versions plus £12

Type 101 1MHz 100KHz 10KHz Crystal Standard £80
Type 103 Off/Air Standard £78

SUPPLIERS TO: Ministry of Defence, G.P.O., B.B.C. Government Dept., Crystal Manufacturers and Electronic Laboratories world-wide.

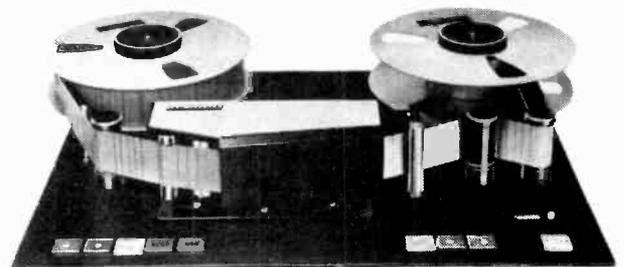


R.C.S. ELECTRONICS
NATIONAL WORKS, BATH ROAD
HOUNSLOW, MIDDX. TW4 7EE
Telephone: 01-672 0933/4

WW — 048 FOR FURTHER DETAILS

NEW FROM *brenell*

MULTITRACK 600 SERIES
2" TAPE TRANSPORT



Price £1,850 + VAT

Also available complete with Electronics

Finance Available

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WW—009 FOR FURTHER DETAILS

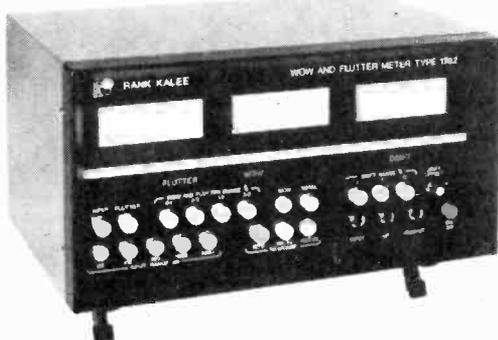
TV Waveform Monitor EV4040...



One of the EV range of TV waveform, vector & picture monitors - ask for details

ELECTRONIC VISUALS LIMITED P.O. Box 16, Staines, Middlesex, TW18 9LF Tel: Staines 56186.

WW-093 FOR FURTHER DETAILS




**The
new
Rank**

WOW & FLUTTER Meter Type

1742

Fully transistorised
for high reliability

Versatile

Meets in every respect all current specifications
for measurement of Wow, Flutter and Drift
on Optical and Magnetic sound recording/reproduction
equipment using film, tape or disc

High accuracy
with crystal controlled oscillator

Simple to use
accepts wide range of input signals with
no manual tuning or adjustment

Two models available:
Type 1742 'A' BS 4847: 1972 DIN 45507
CCIR 409-2 Specifications
Type 1742 'B' BS 1988: 1953 Rank Kalee
Specifications

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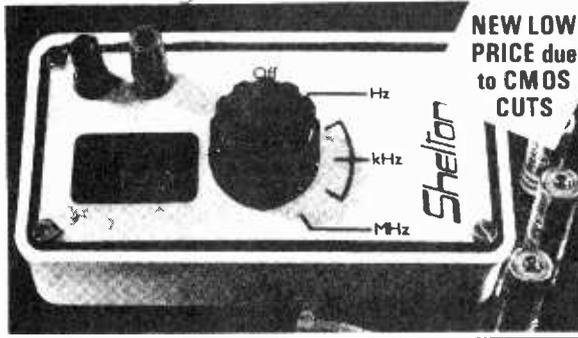
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DC output	20 amps (supply fuse limited)
Power bandwidth	DC to 20 kHz + 1 db—0 db 600 W into 8Ω
Phase response	+ 0 db — 15 db DC — 20 kHz
Slew rate	16 V/μsecond
Damping factor (8Ω)	greater than 400 DC—1 kHz
Hum & noise	120 db below 600 Watts
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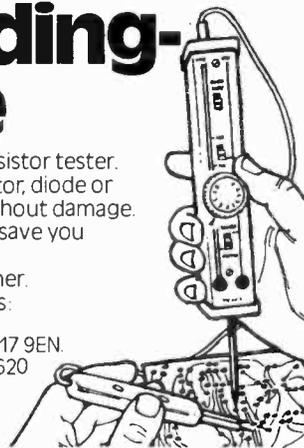
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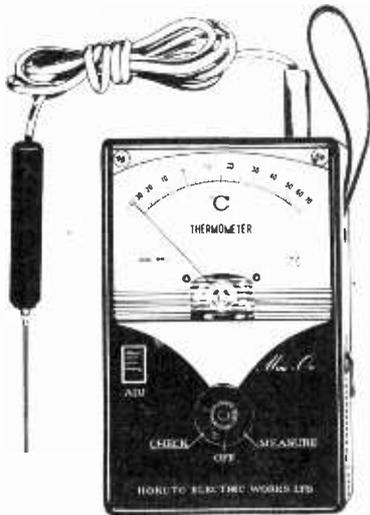
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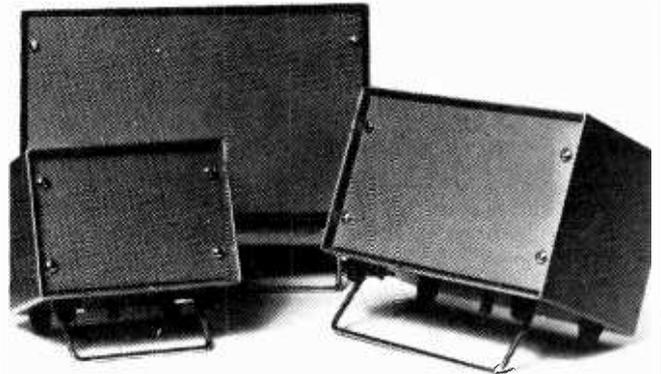
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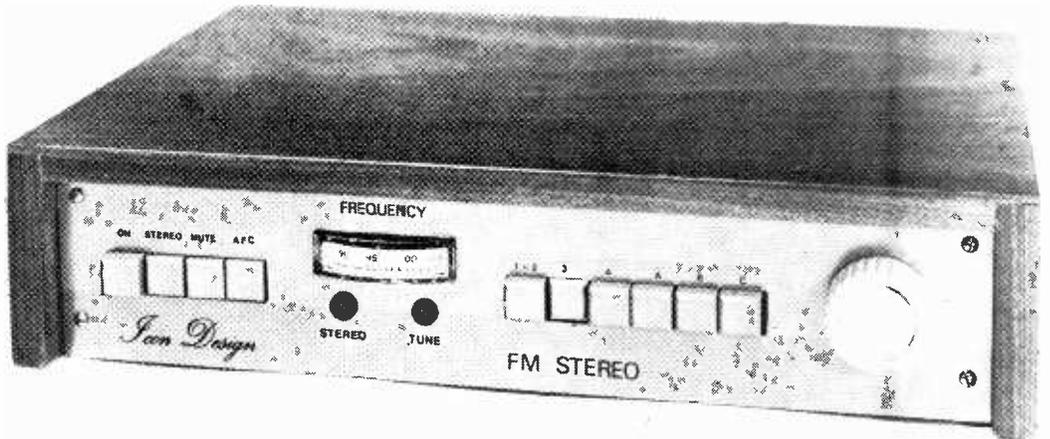
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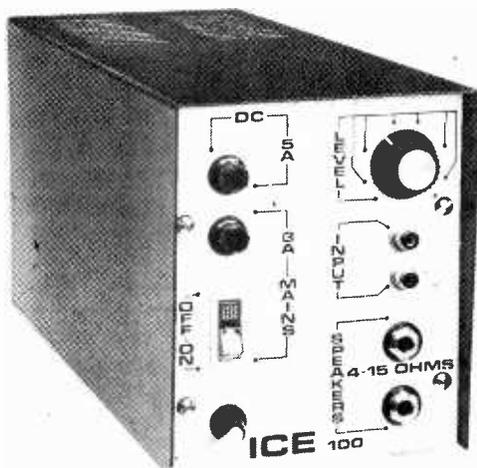
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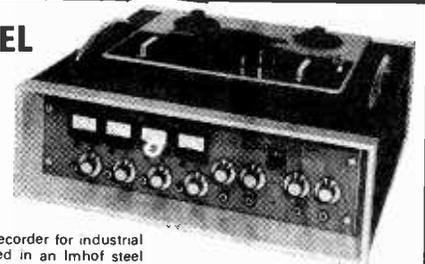
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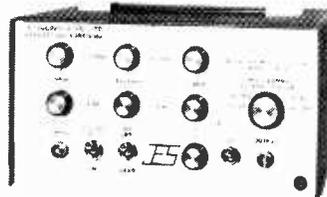
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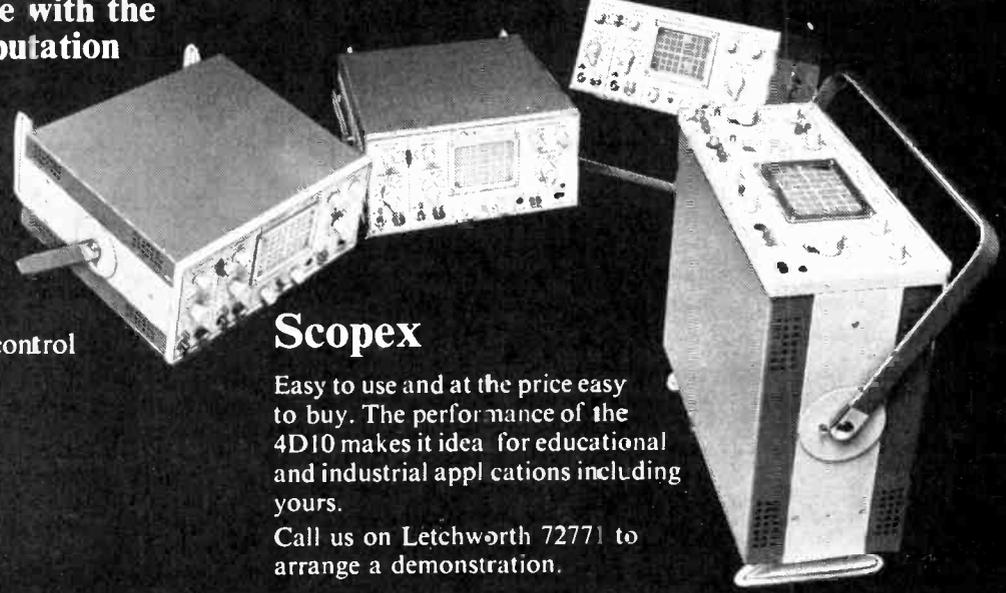
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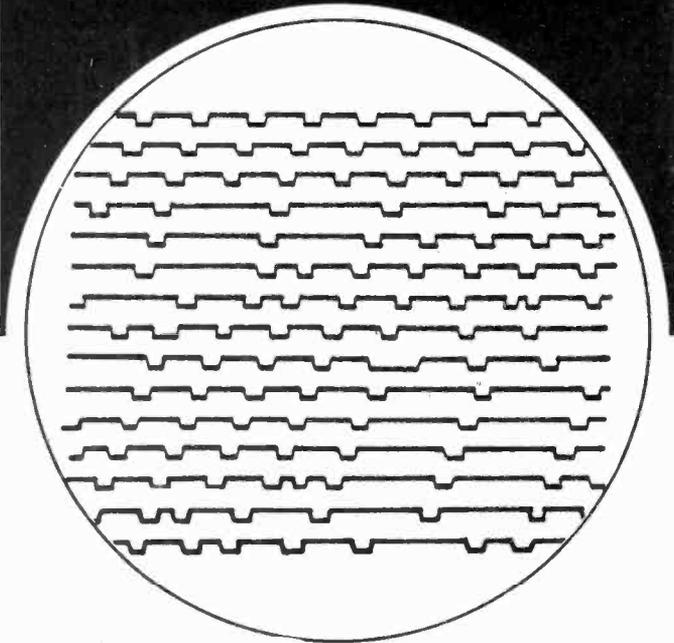
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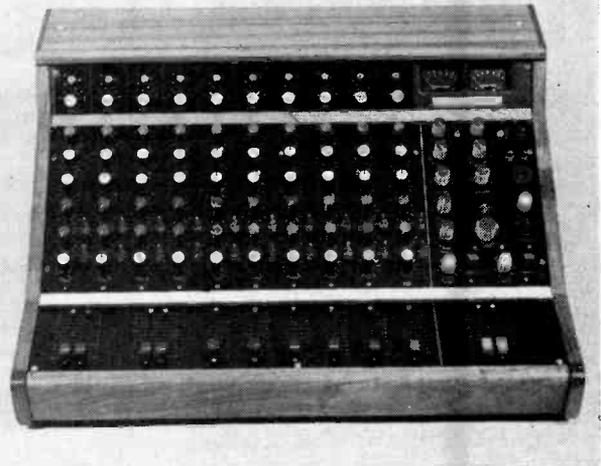
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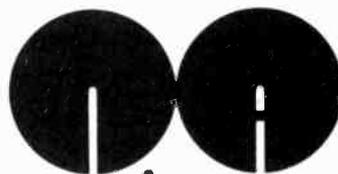
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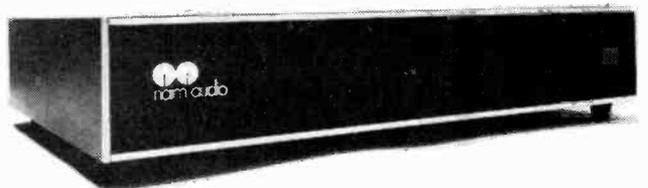
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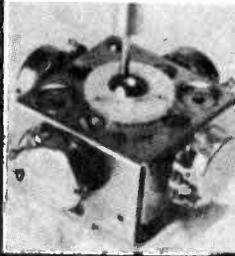
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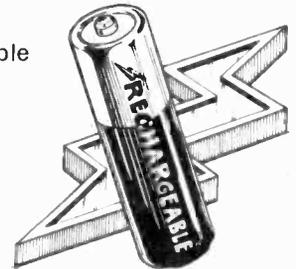
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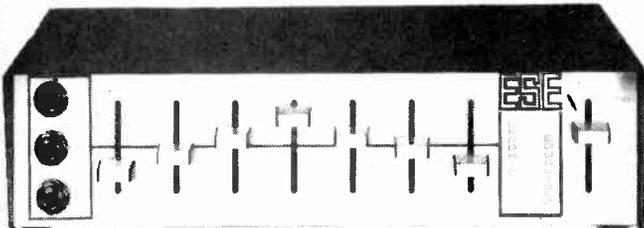
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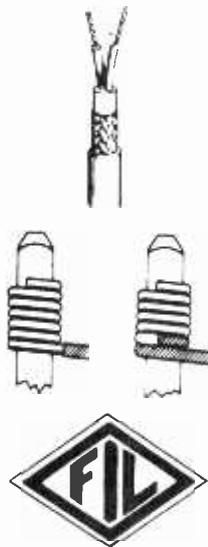
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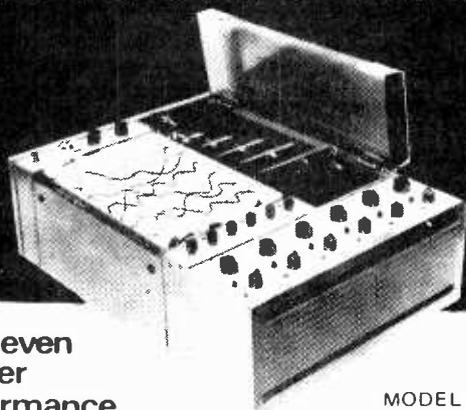
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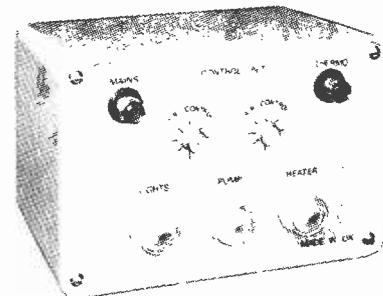
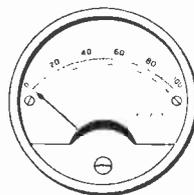


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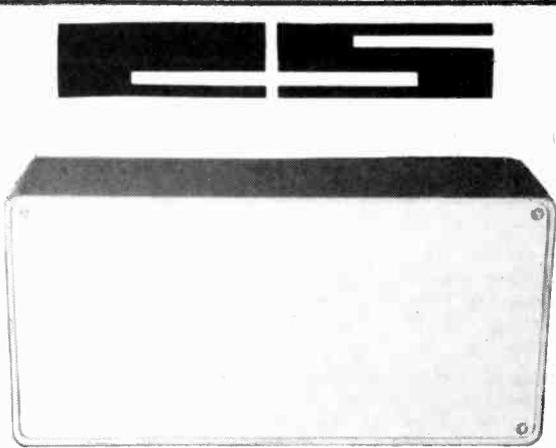
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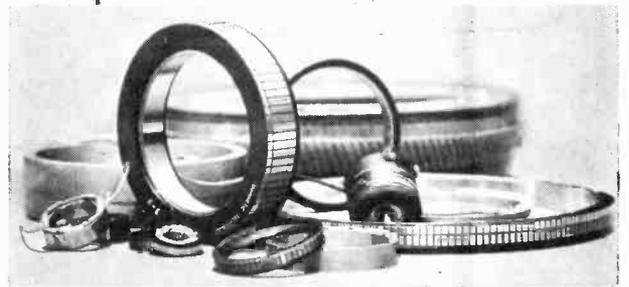
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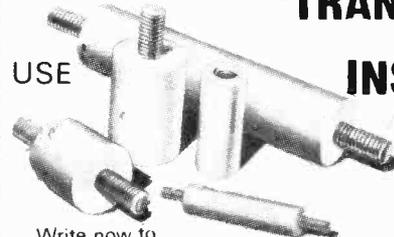


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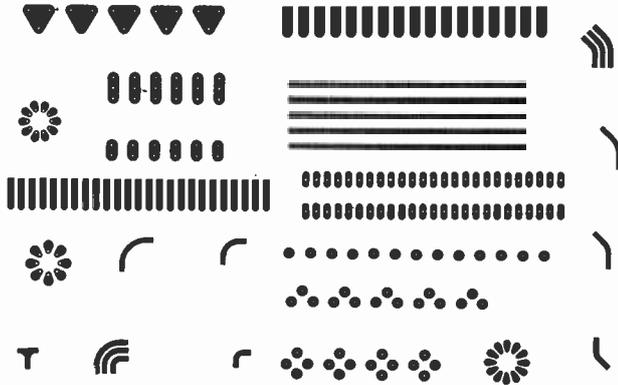
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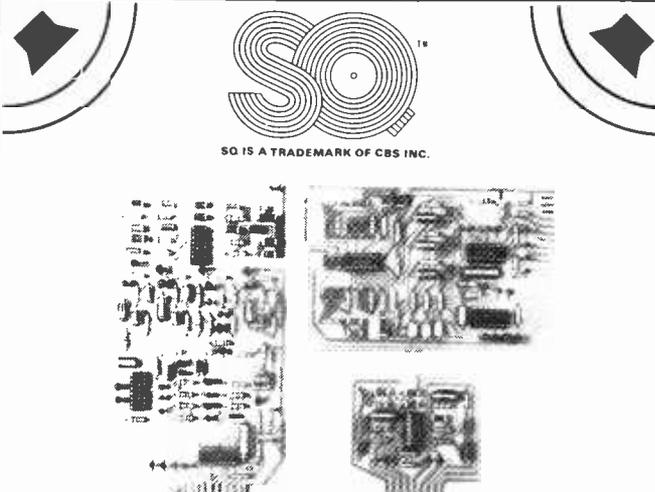
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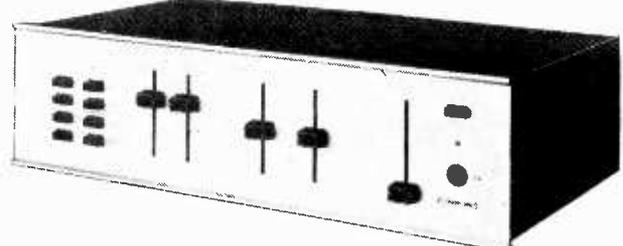
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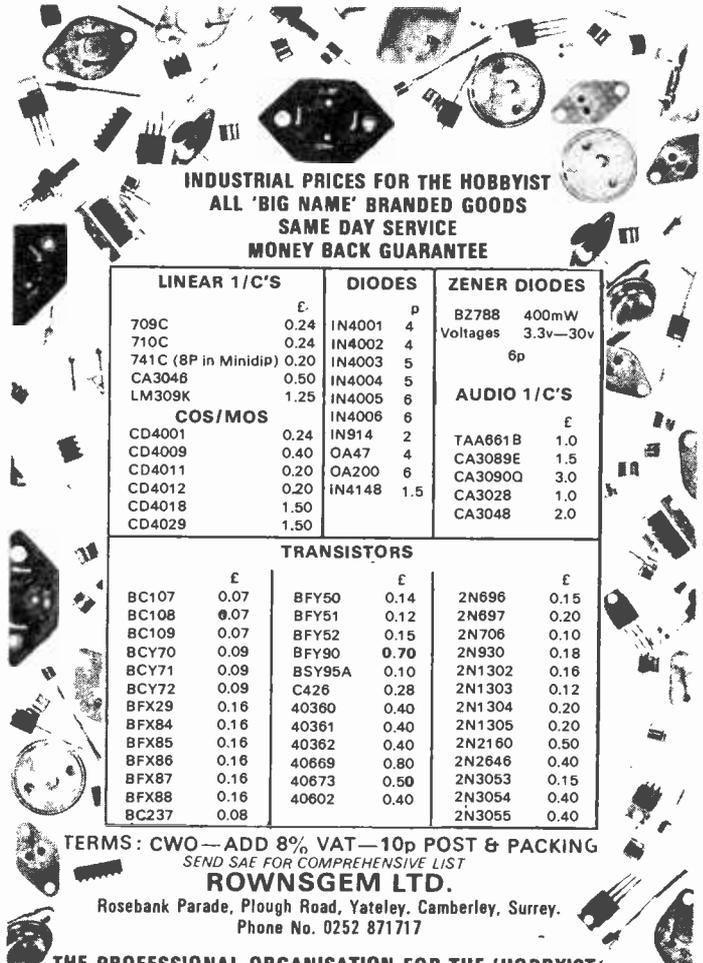
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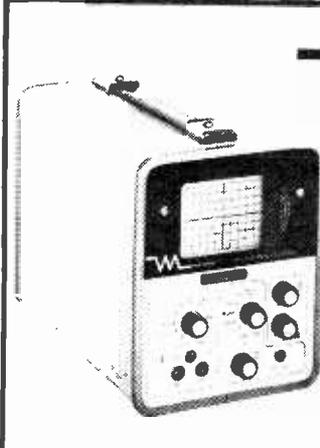
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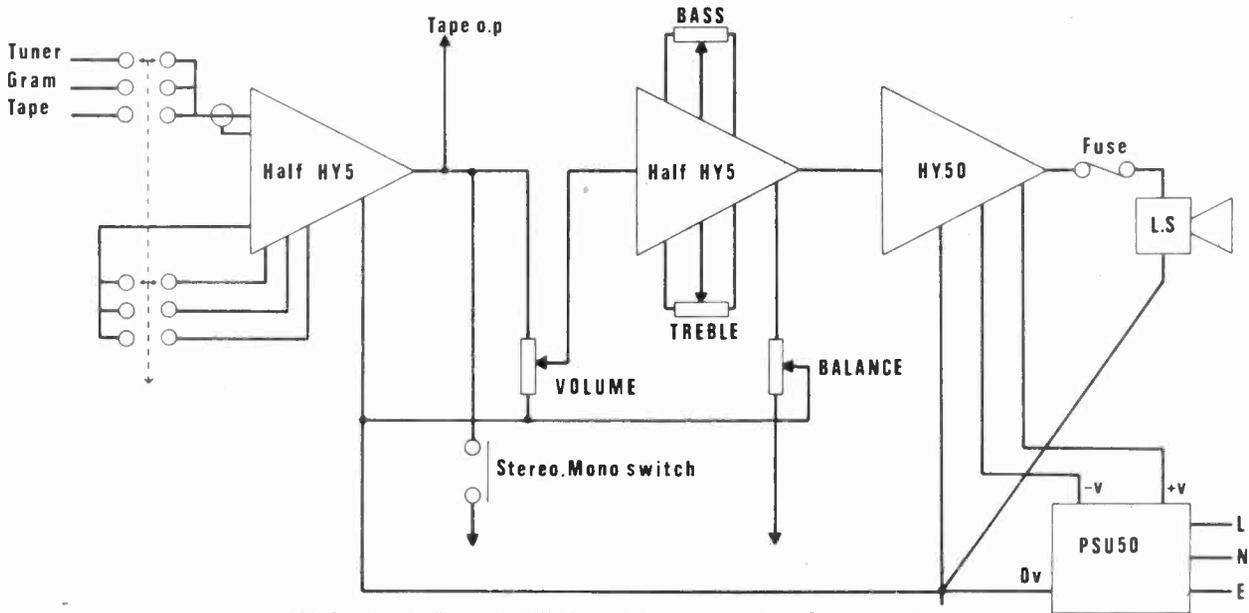
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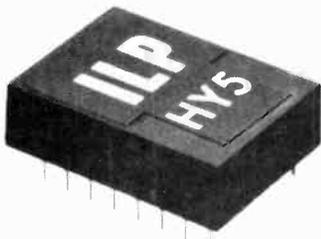
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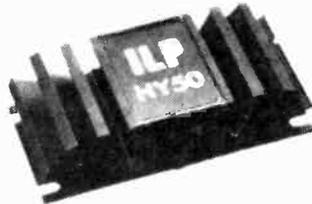
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Input Impedance 47kΩ

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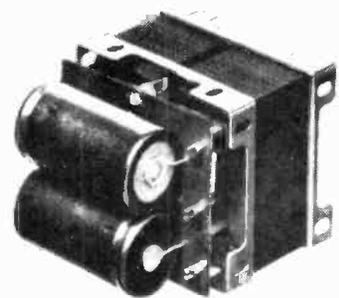
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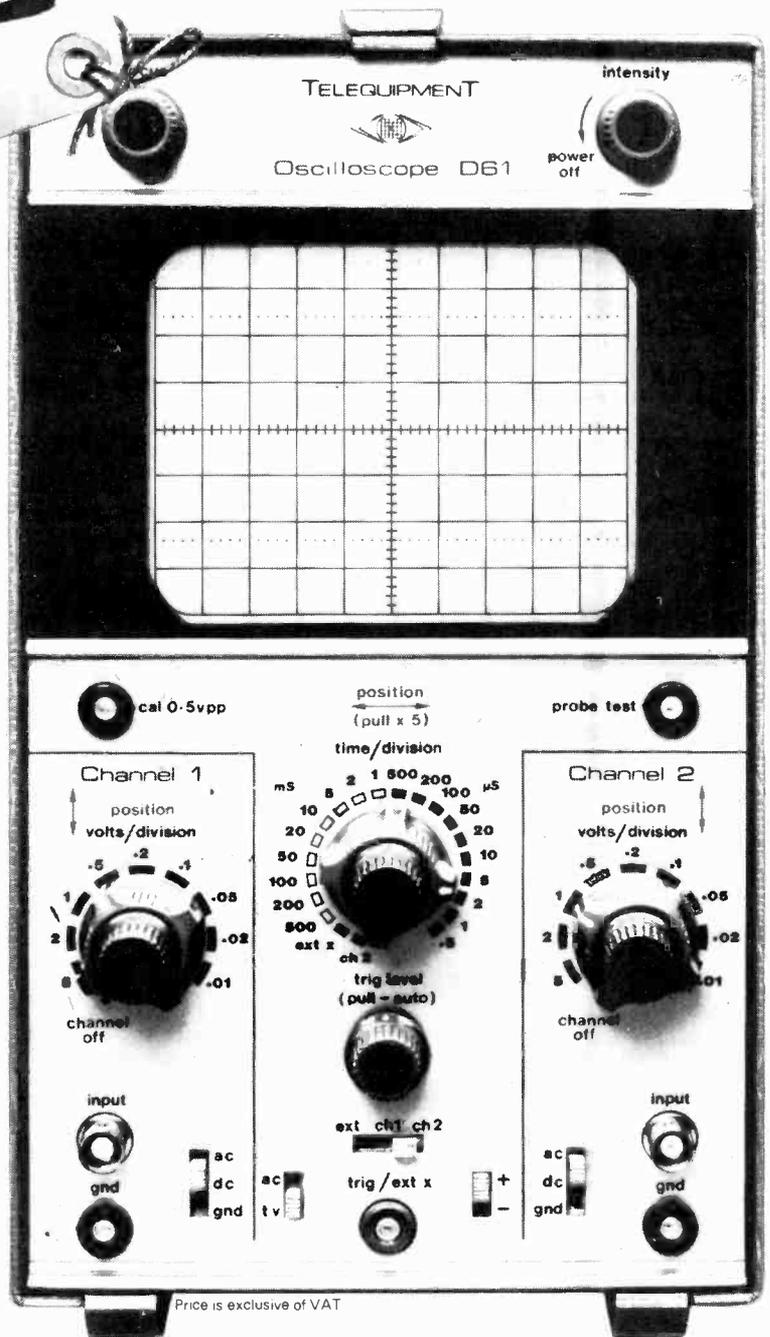
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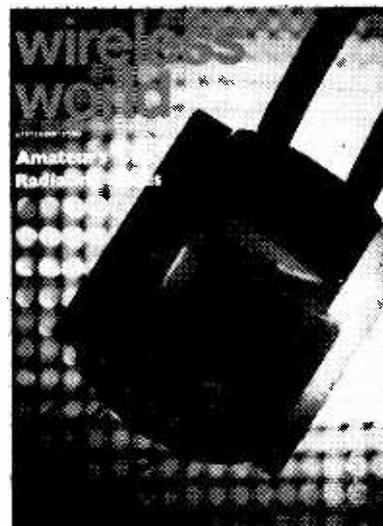
wireless world

Electronics, Television, Radio, Audio

SEPTEMBER 1975 Vol 81 No 1477

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This month's front cover shows the construction of an Audix Ltd opto coupler for conveying signals between circuits which have to be electrically isolated
(Photographer Paul Brierley)

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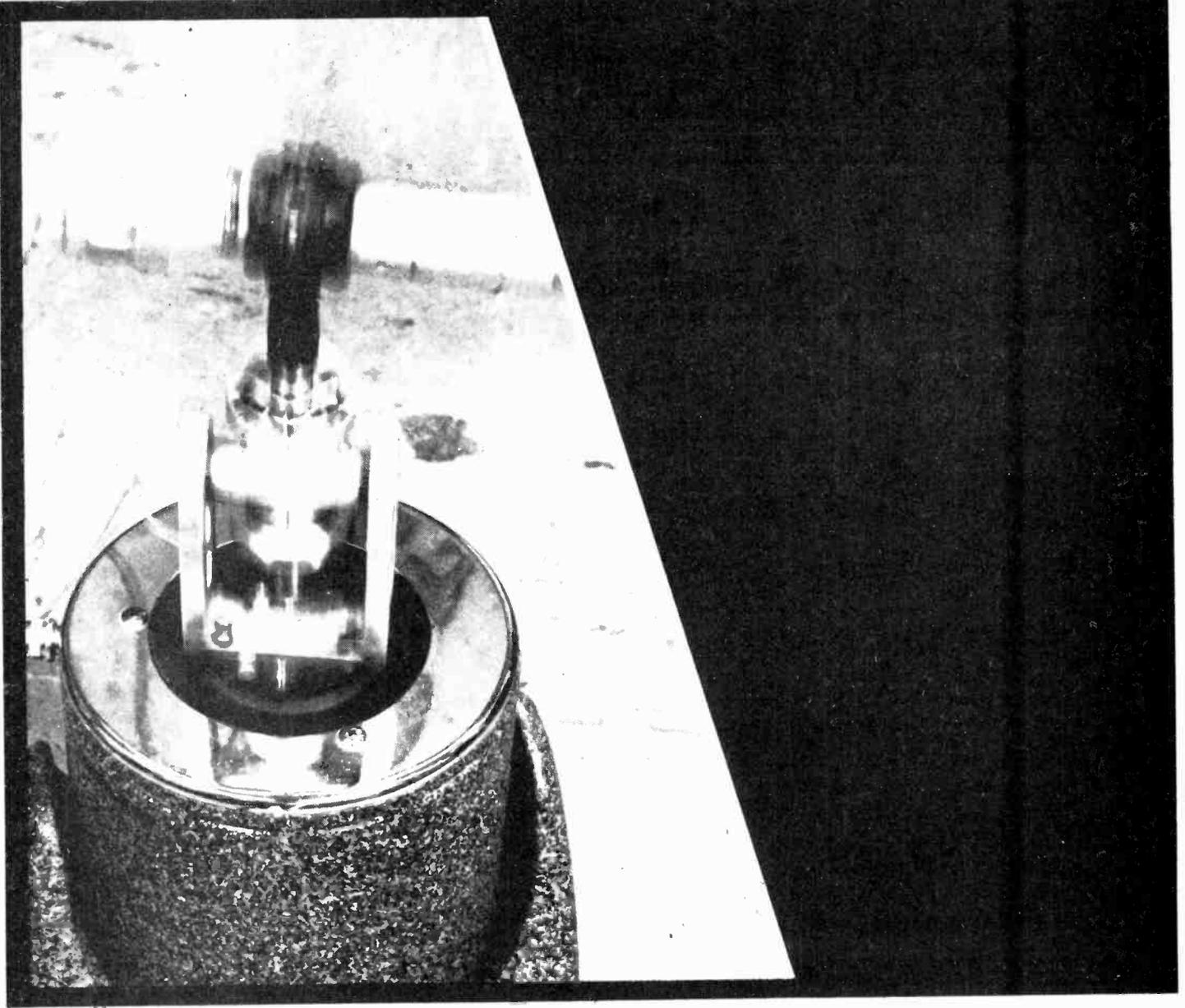
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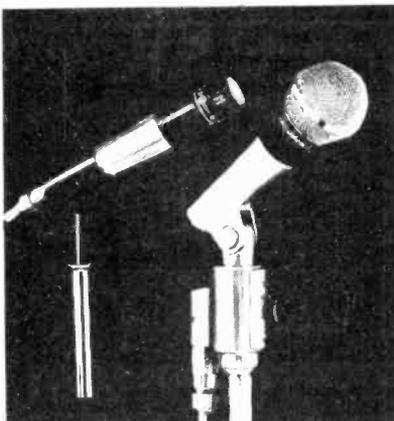
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There are still some rugged individuals around who believe that the consumer should be able to look after himself. *Caveat emptor*, let the buyer beware, and all that. They really think it is healthier for him to protect his own interests when purchasing goods, even at the risk of being swindled occasionally. This is all right when you're buying a quarter kilo of butter or a pair of shoes. But when you are faced with the more complicated and expensive products of our technological society — hi-fi equipment, colour television sets, electronic calculators, electronic watches and the like — it's altogether a different matter. You really need the help of the experts in the collective form of consumer protection associations and their publications. Unfortunately our rugged individuals are supported by many traders who regard "consumerism" (they probably invented this pejorative term) as an unwarranted interference with their right to sell to people any kind of rubbish they can be gulled into buying.

If the consumer protection movement were left on a national basis, things would have probably bumbled on unsatisfactorily for the British consumer for a long time. But now, with our membership of the Common Market confirmed by the referendum, there is a new twist to the situation. European consumerism arrives with the force of law from Brussels. The Treaty of Rome says there shall be no technical barriers to free trade among the Member States, and the EEC has been busy framing common standards to overcome these technical barriers. In our own field much of this is based on the work of the long established International Electrotechnical Commission. Whereas the main object of common, or "harmonized", standards is to permit free trade, some will also have the effect, because backed by European law, of enforcing consumer protection. One such standard is the so-called "Low Voltage Directive" for the safety of electrical equipment, which is discussed in some detail in this issue.

Such European initiatives are to be welcomed. One cannot help having doubts, however, about the length of time they are bound to take. EEC standardisation on electrical safety began with a conference in 1966. If such relatively straightforward questions as the practices likely to cause fire or electrocution are going to take so long, what will happen with standardisation which depends on more subjective criteria such as the quality of sound reproduction or television pictures? By the time the cumbersome process of decision making has reached a conclusion the technology will have moved on and subjective standards will have changed. But the heart of the EEC seems to be in the right place, and if the motivation is strong enough some good for the consumer will come out of it.

Radiating cables in buildings and city streets

An investigation of radiating properties for localized radio coverage

by R. Johannessen & P. K. Blair

Standard Telecommunication Laboratories

With the growing needs for radio-communication, radiating cables are finding increased acceptance in many different areas as a convenient interface between radio base stations and space. The usefulness lies in the degree of control of the coverage given to the designer in that good radio communication is achieved between a base station and mobiles in the vicinity of the cable whereas other localities have a restricted field strength. Thus by using radiating cables and locating them where communication is desired, the frequency spectrum pollution is greatly reduced compared to the case when conventional antennas are employed at the base station.

This article describes two uses of radiating cables in which this feature is important. The first is in an indoor application where the object is to achieve good communication within a building coupled with minimal external leakage. In the second case, a possible

use for such cables in city streets is considered.

Indoor application

In order to obtain a qualitative idea of the performance of such installations, extensive measurements have been carried out at the building complex belonging to STL in Harlow and it is these measurements which are reported on in this section.

The laboratories are located in flat countryside immediately east of the A11 Harlow bypass. Between the A11 and the laboratories is a cluster of trees. Towards the south and east of the laboratories the aspect is generally open. Fig.1 shows a plan view of the layout of the buildings. Each building is known by a reference letter shown in

Fig.1, thus the four buildings facing south are known as U,S,E and D.

The main features of the site are as follows:-

Four similar blocks U,S,E and D are set out in a straight line, each block having two floors. At the east end is C block having four floors. These five blocks are characterized by having a central corridor with laboratories and offices on both sides. Corridor and labs are separated by walls which, over 90% of their height, have a solid metal construction, the upper 10% being glass. Z block has a metal/glass wall running east west such that the northern part of the building — about ¼ of total floor area — is offices, the rest being laboratory in generally open plan arrangement. The north and south walls of blocks U, S, E, D and Z plus the east wall of block C are about 50% glass.

Three cables are located in the laboratories. They all start between D and E blocks. One cable runs through E,

Fig. 1. Layout of building complex and summary of results.

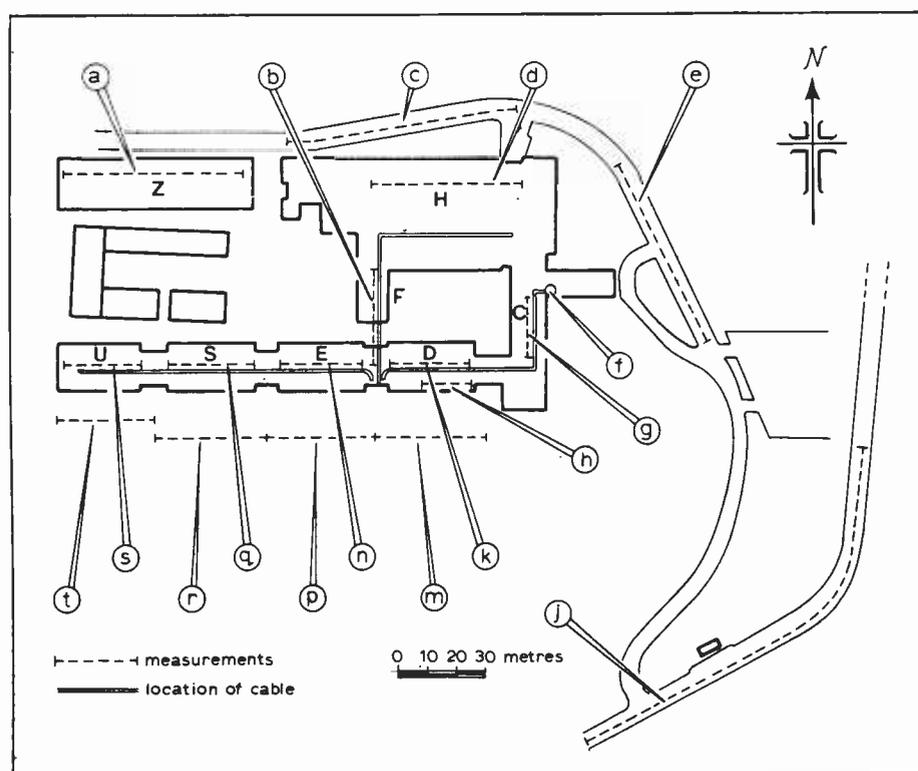


Table 1

Reference in Fig. 1	Floor level	Coupling	Standard devn.
a	2	109	3.4
	1	129	5.6
b	2	66	4.1
	1	56	4.9
c	1	116	3.6
d	1	93	4.1
e	1	114	5.1
f	X	87	6.4
g	4	87	4.7
	3	79	4.6
h	2	71	3.3
	1	58	5.4
j	1	79	5.7
	1	110	6.0
k	2	66	3.2
	1	49	4.8
m	1	87	3.9
n	2	52	4.0
	1	44	3.6
p	1	85	2.8
q	2	60	3.6
	1	47	3.8
r	1	89	3.5
s	2	78	4.4
	1	54	3.4
t	1	96	3.6

S and U blocks between ground floor and the 1st floor. The second cable runs along D and C block between ground and 1st floor level and then runs up the elevator shaft at the end of C block. The third cable runs along F and H blocks between ground and 1st floor. A cross section of buildings U, S, E, D would show that the cable is positioned above the ground floor suspended ceiling. Thus between the cable in S block and the front of the building is first the metal/glass wall and then the outside wall.

The cables used were all of the long continuous slot type with an opening of 180°. The nominal coupling of the cable at 450MHz is 60-70dB for a range from the cable axes of 1 to 5 metres and the insertion loss is in the region of 73dB/km at the same frequency. Both coupling and insertion loss are the values measured with the cable in a nominal free space environment.

The three cables were connected together so that a reasonable impedance match existed. The combined cables were energised at 454MHz (u.h.f. band) using a Starphone mobile radio transmitter thoroughly screened.

The signal levels at different points inside and outside the buildings were measured by a series of runs. Each run is shown in Fig.1 as a broken line along which a receiver was moved. For each such run some 100-200 spot measurements were taken of the received signal level from a vertical $\lambda/4$ dipole connected to the receiver. The results for each run are shown in Table 1. The first number for each run is the floor number such that 1 represents the ground floor. The second figure is the difference in dB between the signal power received and that flowing out of the transmitter. The mean value for the run is the one recorded. The third figure is the standard deviation for all the measurements for that run.

Results commentary

Note that the signal level drops as one moves along the cable from block E through block S and to block U. Similarly from block D to block C. This is mainly due to the insertion loss of the cable. Also the signal level is strongest on the ground floor and weaker one floor up. This is due to there being thin false ceiling tiles between ground floor and the cable, whereas between cable and the top floor is a layer of concrete which attenuates the signal. In the case of C block the two upper floors are in general far from the cable. The signal level there is due partly to that picked up from C block ground floor, partly from the lift shaft of C block and partly from the cable running into F block. In the case of Z block the signal level on the top floor is stronger than on the ground floor. This is probably due to the shadowing caused by the low buildings positioned between U block and Z block.

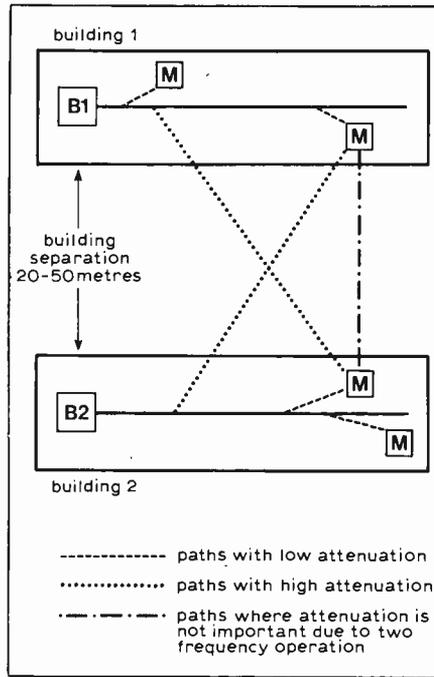
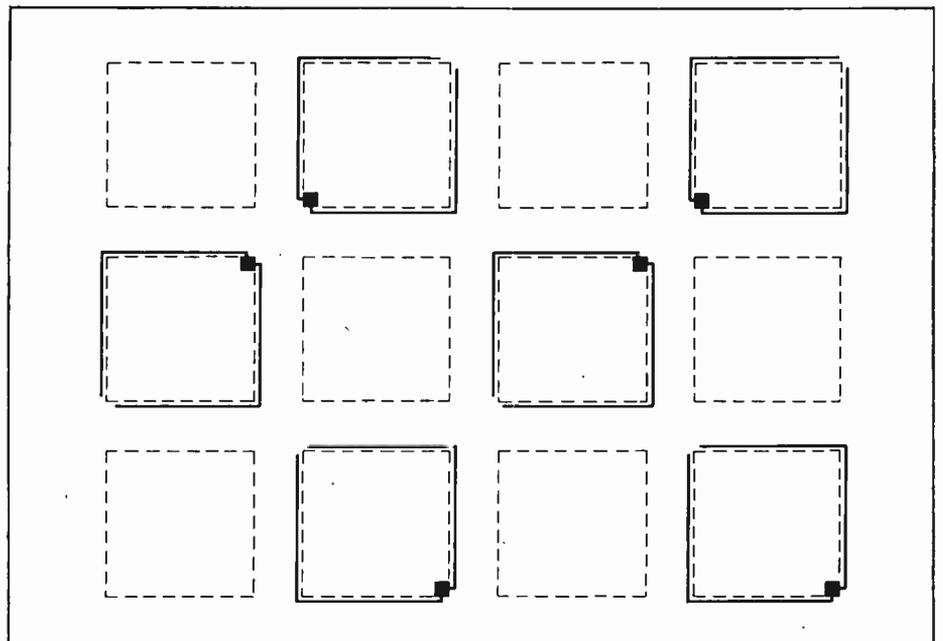


Fig. 2. Mechanism of co-channel working. "M" indicates mobiles.

A very large difference is noted between ground floor of F block and ground floor H block. This is due to two factors; first, measurements in H block had to be carried out much further from the cable than was possible in F block and second, a solid brick wall was positioned between the cable position in H block and the place of the measurement. Along the front of the building it will be noted that there is in the order of 40dB difference between outside measurements and those on the ground floor immediately under the cable. The signal level on the surrounding roads

Fig. 3. Arrangement of radiating cables in small cell system.



has in all cases a mean value at least 100dB below the level fed into the cable.

It will also be noted that the coupling on the ground floor in buildings U, S, E, D, C and F is closer than the nominal coupling for the cable. This is probably due to the cable coupling into pipes and other cables located in the same ceiling ducting as is the radiating cable.

Fig. 2 indicates how it is now possible for mobiles in building 1 to talk to base station B1 via its cable, while the mobiles in building 2 talk on the same frequency to base station B2 via its cable.

An indication of the degree of protection between two buildings can be given by considering Z block compared to S block. In Z block upstairs the received signal level to 95% probability is $109 - 1.96 \times 3.4$ which is 102.3dB below transmitted signal level. In S block upstairs the 95% probability for wanted signal is $60 \times 1.96 \times 3.6$ which is 67.1dB below transmitted signal or a difference of 35.2dB from that in Z block.

Cables for city streets

Commenting on the increasing demand for radio spectrum, paralleled by the population explosion, Hardeman¹ says that in the US, pockets of excessive crowding occur in just about all bands under 10GHz and new technology is one of the means required to relieve the pressure. Land mobile users in the US have saturated the v.h.f. and u.h.f. bands and are now searching for the most efficient methods to apportion a new band at 900MHz. Japan has an equal problem both at 150 and 460MHz and Linney² has outlined the steps the UK Post Office are taking to meet the growing demand for their Radiophone service. The problem is the same - where to find frequencies to satisfy a growing number of users.

An approach which has been proposed for use in the US cities is the cellular approach^{1,3} which takes advan-

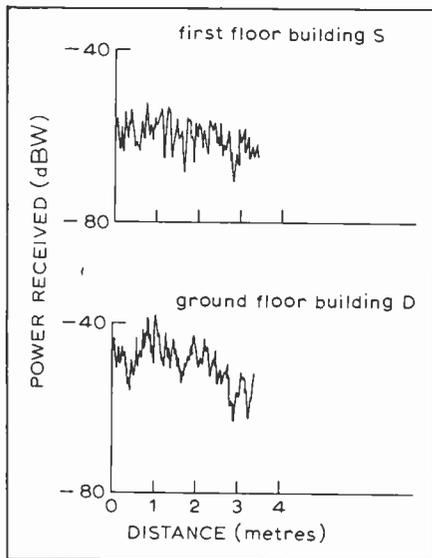
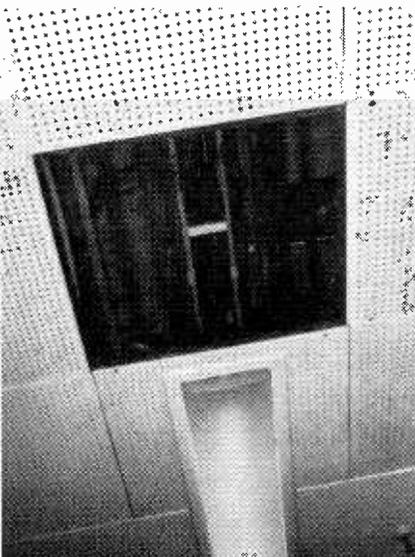


Fig. 4. Signal level variations measured in buildings S and D. The power received (vertical scale) is at the terminals of $\lambda/2$ dipole for 1W into the cable system.



Three types of radiating cable made by STC Cable Division.



Ceiling of E1 showing how the cable is laid without any complicated fixings and next to a variety of other pipes and cables.

tage of the large geographical distances between co-channel mobiles. It is claimed that potentially, the cellular approach requires much less spectrum to provide a given quality of service to the user. The size of the cells determines how efficiently the radio spectrum is used — the smaller the cells the more times the frequency can be repeated without mutual interference between users. But the smaller the cell, the more complicated and costly must be the organization of the system including position finding of the mobile, assigning of frequency and switching of frequency as the mobile moves from one to a neighbouring cell.

Cables for the cell

One problem with the cellular approach is the extent to which a message to/from one cell spills into a neighbouring cell. This is aggravated by the need to overcome building shadowing with higher transmitter power than that required for free space propagation. Thus in order to have sufficient signal strength in all streets within one cell, there is likely to be excessive signal strength in some parts of the neighbouring cells.

Radiating cables appear to offer a natural solution to this since the signal level drops off rapidly as one moves beyond the cable ends⁶. Furthermore if the radiating cables are located along the streets, effectively distributing the antenna where coverage is desired, there will only be minimal shadowing by the buildings.

In Fig. 3 a possible approach is shown in a city where the streets can form a regular pattern. The broken lines represent the building outlines. The shaded blocks are transceivers connected to one or more radiating cables. The adoption of centre or end feeding for the cable will depend upon building geometry and frequency used. Assuming the base of a building is 100 metres square and an end fed arrangement is used, the cable would be around 400 metres long corresponding to an insertion loss of typically 12dB at 160MHz or 33dB at 900MHz.

An audio connection will be required between the transceivers working the cables and a central exchange. Possibly ordinary telephone line could fulfil this function.

Assuming the radiating cables are located on the external walls of buildings one can expect a coupling loss of about 90dB. The precise loss is a function of cable type, fixing method employed and the degree to which the surroundings re-radiate. Four hundred metres of cable operating at 160MHz can be expected to have an insertion loss of about 12dB. Man made interference levels in urban areas are known to be high; however, it will be recalled that whereas the insertion loss of a cable increases with increasing frequency, the ignition interference level falls.

Walker⁴ suggests a drop in interference level of 20dB when the frequency is increased from 150MHz to 900MHz.

Vehicle location

Common to all small cell systems is the need for vehicle location and following. The control system must know where a vehicle is and as it moves towards the boundary with another cell another frequency must be in readiness or the same frequency in the neighbouring cell must be cleared. The smaller each cell is, the greater becomes the requirement for vehicle following; thus more rearrangement of channels may become necessary during one particular radio-call. McClure⁵ has outlined some of the control functions and formats which could be used in a free radiating small cell system. The radiating cable provides a medium for position location and a variety of methods could be considered as candidates for study.

It will be appreciated that there are very many further aspects to this proposed scheme which require further study and careful analysis. For example, the interface and interaction with the telephone switching system, the channelization scheme and associated controls, position finding methods for operation with radiating cable systems, the integration of the system with other "free radio" schemes and, last but not least, its cost effectiveness.

This article was originally presented as a paper at the Communications 74 conference on radio and data communications held in Brighton.

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Electrical safety, standards and the law

The background requirements for electrical safety

by Basil Lane

Assistant Editor, Wireless World

September 1, 1975 sees the introduction of a new regulation under the umbrella of the Consumer Protection Act. It results from the promulgation of the "Low Voltage Directive" from the Common Market Commission and brings within the scope of trading standards officials the new question of the electrical safety of a wide range of electrical appliances, not previously covered by Home Office regulations. This article examines the scope and the background of the law, methods of identifying safe products and the impact of the technical requirements upon manufacturers of electrical equipment.

Since our recent referendum on the Common Market and its affirmative outcome, many of the political, commercial and trading problems under discussion in Community committees become of more direct interest to the UK public at large. Of particular interest to designers, manufacturers and distributors of electrical and electronic equipment are the new safety regulations due to come into force on September 1, 1975.

Treaty of Rome — a background

In signing the Treaty of Rome, the United Kingdom agreed to become subject to a number of basic rules governing membership of the European Economic Community (EEC). These rules are contained within the Articles, a number of which relate to trading within the Community. For example an extract from Article 3 says that there should be moves towards "the elimination between Member States of Customs duties and of quantitative restrictions on the import and export of goods and of all other measures having an equivalent effect." The first two speak for themselves, but the second requires some further qualification. This is provided in Article 100 of the Treaty of Rome, which provides for the removal of barriers to trade which are based on technical grounds.

Since one of the principal barriers to trade between countries is the often conflicting electrical and other safety requirements, this was clearly an area that the EEC Council had to examine. The results of their deliberations came on February 19, 1973 in the form of an EEC Council directive to the Member States. Commonly called the Low Voltage Directive, it has the somewhat more specific subtitle of "On the

harmonisation of the laws of the Member States relating to electrical equipment designed for use within certain voltage limits."

Within this directive are 14 Articles which in substance make the following points to the governments of the Member States. The comment is made that at the moment some of the Member States require electrical goods to meet specific requirements before they can be offered for sale. In others, no such regulations exist. The directive therefore requires that the Member States should move towards a common form of safety legislation based on "provisions or standards already laid down by other international bodies or by one of the bodies which establish harmonised standards."

This means the recommendations of the International Electrotechnical Commission (IEC) or the standards agreed by the *Comité Européen de Normalisation Electrique* (CENELEC), a section of the larger body CEN. The CENELEC standards on safety are adopted from another European organisation CEE, which deals only with electrical safety matters.

Since no harmonised standard exists as yet, the directive acknowledges the fact by laying down certain requirements which were supposed to be implemented by law, within each of the Member States within 18 months of the date of the directive. There was one exception, that of Denmark, where existing legislation was rather more complex, and the Council felt that they should be given up to five years to fall into line.

As yet, the British Government has not implemented this directive in law, but is about to do so (at the time of writing) and so it is impossible to quote

the letter of the law as it will appear. For this reason, and since I am assured that it contains no more than appears in the directive (though couched in different terms), a closer examination of the Articles of the directive is desirable.

The first Article says that the directive deals with all electrical equipment designed for use with a voltage rating of 50V a.c. to 1kV a.c. and 75V d.c. to 1.5kV d.c. The second Article says that laws must be made to ensure that only goods "constructed in accordance with good engineering practice in safety matters" and that it should not, when properly used, endanger humans, domestic animals or property. The main requirements of this Article are listed under Annex 1 of the directive and include proper marking of the goods with the manufacturer's trade mark or brand name, proper assembly of components for safety, proper design to avoid electrical hazards in normal use; temperatures, arcs or radiations likely to cause harm should not be produced, proper protection against non-electrical hazards should be provided for and standards of insulation should be adequate for foreseeable conditions

In addition mechanical requirements should be of such a standard that no hazards are caused by external "non-mechanical influences" or by overload conditions.

Article 3 says that, subject to the provisos of subsequent Articles, if the goods meet the previous requirements that free trade within the Member States shall be allowed and not prevented on the grounds of safety. Article 4 prevents the application of stricter regulations than those listed by electricity companies before permitting connection to the supplies. Article 5 says that harmonised standards should be

used to determine compliance with Article 2 and defines a harmonised standard as being one which has been drawn up by common agreement of the Member States by a competent authority. These standards should be flexible to allow for upgrading to suit technical advances.

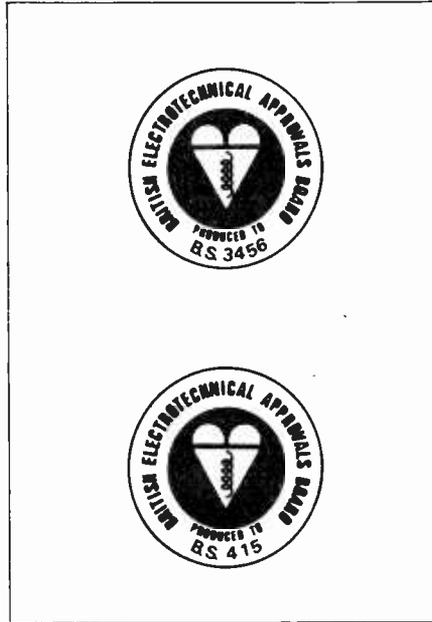
Article 6 is important, since it relates to the situation where there are no harmonised standards and says that in that case, the authorities should regard equipment which complies with the recommendations of the IEC, or the similar document prepared by CEE, as being of the appropriate standards of safety. If there are any objections to these standards, then they have to be made within three months of the date of the directive. (The UK has entered such an objection stating that the recommendations of BS415:1972 are of a higher standard of safety and should therefore be adopted particularly in respect of sound and vision equipment.)

Article 7 provides also for the situation where no harmonised standard exists and says that where the safety regulations applying to goods manufactured in other Member States are equal to, or higher than, ours, then we should regard such products as being suitable for free trade and within our own standards of safety.

Article 8 is both complex and far reaching, since it admits the possibility of the harmonised, or accepted, standards being inadequate in the light of technical development. Essentially it says that where a piece of electrical equipment does not conform to the recognised standard, but the manufacturer claims that it meets the requirements of Article 2 (that it is safe), then he may submit a report to an appeal body nominated, presumably by the Member State, for further consideration.

Article 9 deals with the situation where a product is banned from sale or "its free movement is impeded". In such a case the Member State must inform all the others in the EEC that it has done so and then go on to explain why. If an objection is raised by any of the Member States, they must then all go into a huddle and consult. If after three months still no agreement is reached, then the case has to go to one of the appeals bodies referred to in Article 8, but which must be outside the territory of the Member States concerned and not previously concerned with consultations on that case. That appeal body will then make a ruling based on Article 2 only (not with reference to particular standards). After the ruling has been given there is a period of grace of one month for further objections and then the Commission (the European administrative body) will pronounce an opinion or recommendation.

Article 10 deals with safety marks and certification of products. Conformity (and thus free trade) with the directive can be indicated by a safety trade mark



The BEAB safety marks: top BS3456 for "white goods"; bottom BS415 for "brown goods".

or by a certificate. There is also the third alternative, intended primarily, though not exclusively, for industrial equipment, where the manufacturer may make a declaration of conformity. Marks and certificates have to be established by common agreement, by specified bodies appointed to issue them.

Article 11 says that information shall be circulated to all the others by each Member State about the appeals bodies, the safety marking or certifying bodies, the standards bodies working towards harmonisation and who publishes the harmonised version of the standard in each Member State.

Three other Articles conclude the directive, but they are not of immediate interest here. As far as our own regulation is concerned, the substance will be drawn from Article 2 of the directive and the associated annex mentioned above.

Law-making

No Act of Parliament is necessary to bring this regulation within the scope of the law since powers already exist to permit such additional regulations to be made under the Consumer Protection Act. Essentially the document containing the regulation is "laid" with the Minister and under what is known as the "negative resolution procedure" can bypass Parliamentary debate. Unless, that is, a motion is tabled either for its amendment or annulment. In the event that no such motions are tabled, the regulation becomes law from the moment the Minister applies his signature. However, present practice has been to allow a period of grace of a month or two after publication to give all who are likely to be affected time to

become aware of the new regulations. Notes on the enforcement of the new safety regulation will appear later in this article.

Since the Consumer Protection Act applies to all points of sale in the course of a business, the new regulation will also take effect in a like fashion. This means that anyone in the chain of sale of an electrical appliance is liable to prosecution if the goods he is selling do not comply with the requirement of electrical safety. This brings into question the problem of pipeline stocks, and here the period of grace is intended to cope with this situation. It was pointed out by a spokesman for the Department of Prices and Consumer Protection that "It is difficult to defend any general deferment of the regulation since, in effect, it simply requires electrical goods to be safe when used for their intended purpose. It is to be hoped that no equipment which is unsafe in normal use is being offered for sale."

The regulation is, it would seem, as simple as the EEC directive since it does not specifically qualify what is meant by the word "safe". This has already caused considerable argument in committees and meetings and is clearly a point that had to be dealt with by the Department of Prices and Consumer Protection. To provide this information and in accord with the provisions of the Low Voltage Directive, a guidance document has been prepared to be associated with the regulation – but not forming part of it. It will not be part of the law, but is intended as a guidance to enforcement authorities. To quote the DPCP spokesman again "... it is likely that the Courts will take it into account in the event of a prosecution." In defining safety, the guidance document has adopted the approach of the directive and has specified national standards which substantially meet the requirements of the regulation and adds that when a harmonised standard is agreed, this will take over as the defining standard. Again, no copies of the guidance document are available at the time of writing, but it is believed that two of the standards quoted are BS415:1972 and BS3456. The former deals mainly with so-called brown goods (radio, television and audio equipment) whilst BS3456 is mainly concerned with other household appliances.

An advisory committee has been formed to assist in the formulation and updating of this guidance, the members being drawn from industry, trade associations, consumer associations and the DPCP and the BSI.

Safety marking etc.

The Low Voltage Directive mentions the only obligatory aspect of certification or safety marking, by stating that such marks and certificates have to be approved by law (in our case as a trade mark) and the certifying bodies accepted by the Government. The law does not

require goods to be safety marked or certified, nor does it require a manufacturer to make a safety declaration.

But what are the implications if a manufacturer decides not to adopt any of the above procedures? First, it is likely that his products will be subject to scrutiny when exporting within the EEC (thus free trade will be impeded). Secondly his product could be the subject of an early scrutiny by UK enforcement authorities. Thirdly it is unlikely that large department stores or wholesalers will accept his products without proof of compliance with the law, for fear of being prosecuted themselves. Some large stores, such as Littlewoods, already operate a test laboratory of their own to examine goods for electrical or mechanical safety.

What then, are the problems of demonstrating compliance with the law? The best and most "watertight" method is to obtain an approval and safety mark through one of the facilities offered in the UK. Four licencing authorities have been set up here, and the one most concerned with electronic equipment is the British Electrotechnical Approvals Board (BEAB), and its safety marks are illustrated in Fig. 1. These marks have not achieved the status of approval by the Member States, principally because the tests of safety are based on our purely national British Standards. However, under the Low Voltage Directive, moves are being made to provide for eventual acceptance.

The BEAB was formed in 1960 as a non-profit making company limited by guarantee. It is governed by a board of management drawn from the BSI and several trade and industry associations. It is purely an administrative and certifying body and has no laboratory as such. Many tests which it requires are undertaken by the BSI laboratory and ATL. Since the BEAB have produced a 12 page document dealing with the methods of applying for approval, this will not be dealt with here. However, some notes will be given later on how to ensure an efficient processing of an application and test procedures.

The regulations governing the granting of a certificate and mark to the manufacturer are quite tough and subject to a list of rules, called the Certification Trade Mark Regulations. These rules are laid by the Registrar of Trade Marks and very briefly say that nobody can use the safety trade mark of the BEAB without official approval and the granting of a certificate. This is fairly obvious, but what is not so obvious is what may happen if a certificate is not granted, despite a claim by the manufacturer that his product is safe. This suggests, for example, that the product has failed in an area where the appropriate BS may be considered by the manufacturer to be inadequate.

There are two possibilities, depending upon whether the manufacturer

believes the inadequacy exists *before* submitting his goods for approval, or if he disputes the results of tests made already. In the latter case he may submit a report to an appeals committee at the BEAB, consisting of a chairman supplied by BEAB, representatives from BREMA in the case of "brown goods", the electrical supply industry and the BSI. They consider the appeal purely on technical grounds and their decision is final, subject to an appeal made direct to the Registrar of Trade Marks.

Now it should be carefully emphasised that this appeals routine is nothing to do with that mentioned in the Low Voltage Directive, which is more concerned with goods that have been banned from sale by legal process. It is probable, however, that the appeals committee for Marks will also deal with appeals arising out of proceedings instituted by trading standards enforcement officers.

In the case where the manufacturer starts off by believing that the appropriate standard is inadequate, he has the opportunity of claiming so, under what is known as the "innovation clauses" in the two relevant British Standards. In BS 415, this clause comes under para. 3.2 and says "Notwithstanding the requirements specified in this Standard, designs or constructions to which the tests specified do not fully apply but which give an equivalent degree of durable safety may be regarded as complying with the requirements of this Standard, subject to the findings of a special investigation by the approving authority and pending an issue of an amendment or extension to this standard." The same clause is inserted in BS 3546: Part 1: 1974.*

This procedure is of vital importance, since it represents the aspect of the requirements for appeals laid down by the Low Voltage Directive, but in this instance purely at a national level, since we are not, as yet, using harmonised standards for the evaluation of safety. In an excellent description of the activities of the BEAB, Zweigbergk¹ details the special investigation test schedule.

If a manufacturer decides that he wishes to invoke this innovation clause, he says so in his original application for approval and if the appliance is accepted for test it then becomes his responsibility to prepare a draft test schedule for those aspects of safety in that part of his appliance that he considers to be inadequately covered by the appropriate standard. The BEAB then will arrange for the test house to test samples of the appliance, using not only the test schedule prepared by the manufacturer but also tests devised by the test house, which may be added in the light of their considerable experience in this field.

At this stage, a meeting is convened of the technical committee mentioned above and a representative from the manufacturer, to finalise the prepara-

tion of the BEAB "special investigation test schedule." Sometimes the initial tests may suggest that the manufacturer has to make some modification, followed by a further submission of a sample for testing to secure approval. When this is done successfully the manufacturer has to formally ratify the schedule, which is then forwarded to the appropriate BSI technical committee for consideration to be included in the standard. In the meantime, an interim approval and certificate may be granted to the manufacturer, pending publication for consideration to be included in the standard. In the meantime, an interim approval and certificate may be granted to the manufacturer, pending publication of the amendment or addition to the standard. If, however, during the interim period, additional information comes to hand from outside sources (since at the BS stage the schedule is made available to any manufacturer), then the BEAB committee may withdraw approval.

The marks of the BEAB are not the only safety marks to achieve recognition in the UK. Since it was the intention of the Low Voltage Directive that marked goods should have free movement within the EEC, these come under the heading of the "CB" scheme, the CENELEC Protocol Agreement, the \bar{E} mark scheme and the <HAR> mark scheme for cables. These marks and their extent of usage are described in a bulletin from the Technical Help to Exporters (THE) service of the BSI².

Law enforcement

Bearing in mind that it is not a requirement of the law that one should declare compliance with the safety regulations, in any of the forms described above, the enforcement of the law by the trading standards officers will be of immediate interest.

The information printed here is based on information given either over the telephone with appropriate officials, or delivered in speeches by Ministry officials. *It cannot be taken to be an expression of the exact word of the law.*

The enforcement authority is the local Weights and Measures Office (now known as the Trading Standards Office). Its officers are not usually experts in any one subject, such as electrical safety, and thus will usually rely on the advice of qualified experts, or on the offices and test facilities of approvals authorities. They are bound only by the regulation printed in the Consumer Protection Act and do not have to regard the standards mentioned above as being the ultimate arbiter of safety. Even a safety marked appliance can be unsafe, due to inadequate quality control at the factory. Prosecution under the Consumer Protection Act is rare and this largely arises from the wide powers of discretion operated by the trading standards officers. In most cases, if the retailer, wholesaler and

manufacturer react promptly to the comments of the official, by withdrawing from sale or agreeing to modify affected items, no further proceedings will be taken.

The law applies, as I have said, at all stages, even at the design stage, but usually the officer will acknowledge that small retailers may not have the knowledge to be aware of potential safety hazards and so will shift his attentions to the manufacturer. Where a successful prosecution is made, the results of the case have to be submitted to all the Member States, since it could well affect local approval of that same item.

Comment and advice

This part of the article is, as stated, comment and advice offered by the author in the light of the investigations leading to the preparation of this article. It does not necessarily coincide with the official view, but often does reveal little understood aspects of electrical safety and the regulations.

One of the commonest cries heard from those who will be affected by the regulation is, "Why did I not know before? Where can I get such information in the future?" In most instances the answer lies in your trade associations. Each association, provided it is not purely nationalistic in its nature, (i.e. consisting entirely of Japanese product importers) may apply for a place on the appropriate technical committees of the BSI. Since these committees are consulted at various stages during the preparation of regulations, such as the Safety Regulation, prior knowledge may be offered to individual members of associations on a confidential basis. Second, the THE service of the BSI, issues information on standards and relevant legislation, both here and in the other countries of the world. Any manufacturer, exporter (or even some importers!) may join this service. The BSI itself also issues bulletins on national standards, which may be obtained by applying for membership. *Wireless World* itself will also operate a letter enquiry service on this topic (see paragraph at the end of this article) and for the public at large many questions about safety regulations can be answered by the advisory services of the various consumers' associations. The Department of Prices and Consumer Protection also publishes a monthly document called the Consumer Information Bulletin, which may be obtained on request.

So far I have not given any specific information on the goods affected (except in the terms of the Low Voltage Directive), and these are very wide. Such items as industrial plant already have been covered by the law under the old Factories Act and now under the all-embracing Health and Safety of Work Act of 1974 and the original Electrical Regulations of 1908, which still apply. However, under household

appliances, any electrically operated item powered from the supplies listed previously falls under the arm of the new regulation. This even includes such unlikely items as some loudspeakers, since the audio voltages can in some cases exceed the minimum voltage specified! Study of the relevant British Standards is thoroughly recommended, even though the convoluted language and reference to additional standards does not make for easy reading.

As far as safety marking is concerned, most goods designed or manufactured by readers of this magazine come within the purview of the BEAB. It seems that their facilities are being stressed beyond the limits, by the flood of equipment they have now included within their range of approvals (recently all kinds of audio equipment). The BSI laboratories have six test officials working full time and it can take over 14 days to test, say, a television receiver. In a recent call for applications for approval of cassette recorders 174 were received, so it is anyone's guess what the overall situation will be when all other items are called for. Some moves are being made to alleviate the situation, but no information on this has been officially released. In addition, moves are being made to combat complaints from importers that their products are at the back of the queue.

It certainly seems unlikely that there will be very much audio equipment carrying a BEAB label this autumn, not because it has failed the tests, but more likely because it has not even reached the starting post! Delays are being caused by manufacturers not studying all of the BEAB documents, some of the more obvious delays arising from failure to observe the requirement to pay for the test before it is done, and also the supplying of spare parts to replace those damaged during the testing process. If manufacturers were to apply the tests of BS415 (or 3456 where relevant) themselves, prior to submission, and list the results on a report offered with the sample, and then supply the spares they found were needed, with the sample, then the length of time required by the BSI laboratories would be considerably shortened.

The BSI operate an approvals scheme (and Mark) for components and if these are used in the manufacture of an appliance, this too can reduce the time taken for tests. Lists of approved items are available from BEAB (for appliances) and the BSI laboratories (for components). Components and appliances of foreign origin may be supplied for approval and are subject to all the same rules as for British made goods.

Since the marks made are only a type approval, the manufacturer is subject to surveillance visits by BEAB inspectors who have to be satisfied that the standards of production and quality control are high enough to maintain the quality seen in the sample. Finally, failure of the appliance to operate under

the safety tests does not imply failure to comply with the standard as long as it "fails safe".

Enquiries

Professional readers having enquiries about the regulations, standards, or services mentioned in this article, particularly if they are of a technical nature, may write to the Editor putting their points. A number of these may be selected in the future for publication, with an appropriate answer. Readers wishing to make enquiries should clearly mark their envelopes with the words "Safety Regulations". We cannot guarantee to answer all queries with a personal reply, or with specific information, since in some cases it may prove better to refer the enquiry to a more specialized authority.

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* This procedure does not appear in IEC 65 which is proposed as the harmonized standard.

Sixty Years Ago

Low life in Lisbon. The following disgraceful revelation was printed in our September 1915 issue and illustrates the depths to which "wireless" had sunk. One might have known those beastly Huns would be involved.

Portugal

"Three wireless installations fully equipped with Morse apparatus have been found in different parts of Lisbon. One wireless station was discovered on the fourth floor of a house in the town. Five arrests were made of persons, who confessed to having erected three other stations in different localities of Lisbon. They were apprehended by order of the Government. Further information goes to show that the Germans are at the bottom of the matter."

News of the Month

Crystals for calculators

A crystal puller has been introduced which is claimed to be the first designed specifically for the routine production of GaP, a semiconductor used in the manufacture of light emitting diodes. The system incorporates several technical advances giving large crystals with accurately defined shapes. The large scale production of cheap gallium phosphide is expected to lead to its widespread use in calculator and digital watch displays.

The system, called the Melbourn, is 16 feet high and weighs three tons. It grows GaP crystals by the Czochralski method using liquid encapsulation and high gas pressures to prevent the dissociation of the gallium and phosphorus. The system will produce crystals of up to 5kg in weight and three inches in diameter. This is an order of magnitude larger than crystals grown on existing equipment. Crystal diameters can be kept within 1mm of a specified size and this gives high yields of useable material. The Melbourn is the result of a joint three year research and development programme between Metal Research Ltd of Royston, Herts and the National Research Development Corporation.

Bureau of Higher Degrees

Information on the availability of relevant higher degrees in electrical and electronic engineering and physics in the UK is now obtainable from the Bureau of Higher Degrees, a telephone information service, based on a simple retrieval system, which has recently been set up by the Institution of Electrical Engineers in association with the Institute of Physics, the Institution of Electronic and Radio Engineers and the National Electronics Council. The Bureau will provide information on all types of higher awards, especially

industry based degrees and sandwich courses. It could be of particular benefit to training officers and management in educational establishments and industry who are responsible for advising on higher degrees, as well as helping the individual graduate seeking advice on higher degrees.

A telephone call to the appropriate number (01-235 6111 extension 36 for information on physics or 01-240 1871 extension 313 for information on electrical and electronic engineering) will give access to the following information: availability of courses, where these are located, whether Science Research Council or Engineering Industry Training Board assistance is available, details on the courses and a contact within the educational establishment.

Telemetry brings in North Sea oil

Advanced telemetry, control and monitoring equipment will play a major role this year in helping to bring North Sea oil to shore. One of the most recent systems will be a remote control and monitoring scheme which incorporates u.h.f. radio links for use on an offshore drilling and production platform. The equipment will transmit control information and monitor activity over a distance of two kilometres between the drilling and production platform and a single point tanker mooring of about 135,000 tons displacement at which 80,000 ton tankers will load.

In addition to being used to transmit control signals and receive response indications, the u.h.f. radio system will be used as a voice link between the tanker loading module and the 200,000

ton platform. At present, a total of 32 indications will be transmitted over the link and all essential controls and safety devices will operate via the telemetry equipment. One of the most important monitoring roles which the equipment will play is in fault detection. In addition, the telemetry equipment will enable the tanker crew to shut down the pumping operations in the event of a major problem such as a fractured feeder pipe. The supply of this equipment is a contract won by M.L. Engineering (Plymouth) Ltd, from Mobil North Sea.

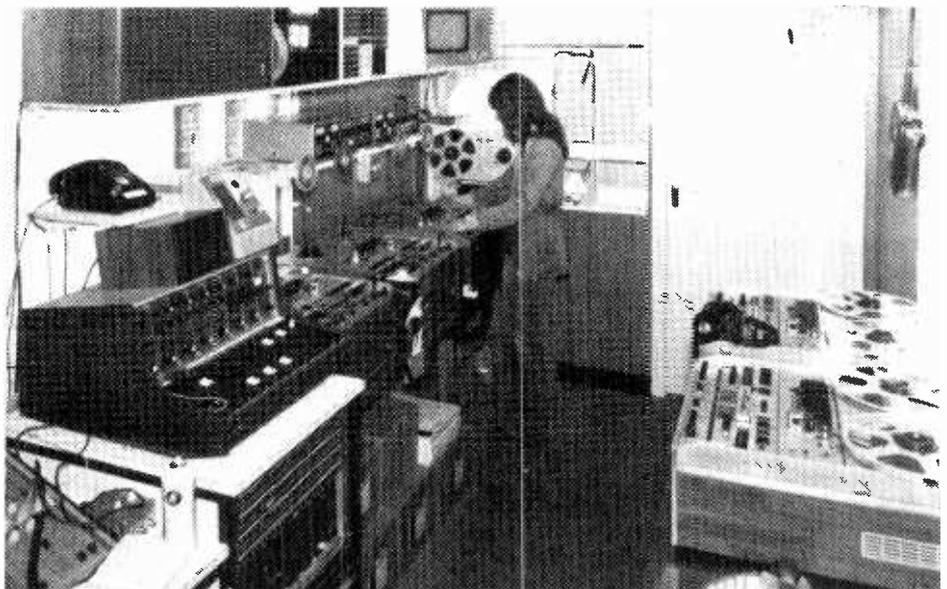
Bouncing ball detector

A method of automatically detecting whether a tennis ball bounces on or near a line and of assisting the umpire in scoring has been developed and patented by a South London inventor, Dr David Supran. The same principle could be applied to several other sports such as squash, football, golf and snooker.

The invention is particularly significant in view of the number of disputes over linesmen's calls at Wimbledon matches and other tennis events.

The cloth cover of the tennis balls used in the invention contains steel fibres similar to those used for the manufacture of special wigs and stockings. These make the cover electrically conductive. In addition, on and adjacent to those lines on a tennis court where key decisions are required suitably coloured tapes can be sited, each bearing 15 parallel channels of flat copper wire. When the special tennis ball bridges any pair of the narrow strips of copper, an electrical circuit is completed and a signal is transmitted to a

Independent Radio News, one of London's broadcasting stations, has taken delivery of six Ferrograph transportable tape recorders for use in its outside broadcast unit at Westminster. Playback is fed out on lines to independent local radio stations, to IRN's central newsroom and to Independent Television News.



Computer Automation minicomputer. The computer is programmed to interpret such a signal as the ball being "out", and a bleep is sounded from a visual display unit next to the umpire. The bleep is audible to umpire, players and spectators alike. If the ball is "in", there is no audible signal but the word "in" is displayed on the v.d.u. screen. In all cases, the umpire can override the system. HE could also control a public scoreboard and other display monitors. A number of tennis courts could be controlled by one minicomputer.

Chart recorder controls furnace

A novel application of an x-y recorder has been developed at the Research Centre of the British Steel Corporation, Motherwell. The recorder controls tempering furnaces to predetermined heating and cooling cycles, in the process of stress-relieving steel samples.

The modular x-y recorder is converted to its controlling role by the addition of a chart drive unit and photo-electric curve follower. The plug-in curve follower module is substituted for one of the amplifier modules, while its light sensing head is mounted in place of the recorder pen. Having decided on the temperature cycle required for a particular sample, the British Steel engineers draw the appropriate profile on the recorder chart for the sensing head to follow. Output voltage level from the curve follower

module is then set so as to be consistent with the change in output from the furnace thermocouple over the required temperature range. The furnace controller is also adjusted to give a temperature reference point. Then, during the process, a temperature comparator compares voltage levels from the thermocouple and the curve follower. The difference determines the operation of the furnace controller and thus the pre-drawn temperature curve is followed precisely. A cold junction reference is inserted between comparator and thermocouple to ensure that control remains constant over wide changes in ambient temperature. The recorder chosen for this application is a Bryans model which has a chart drive unit and photo-electric curve follower as standard accessories.

CCTV in Westminster Abbey

A c.c.t.v. system consisting of three cameras wired to a video selector and with pictures displayed on a 12in monitor has been installed at Westminster Abbey to enable the Abbey organist to see the West Door (where processions form), the Henry VII Chapel (used for weddings) and the Master of Choristers in the choir. A second system for sound reinforcement has also been completed. All the engineering with the exception of the loudspeakers has been carried out by Pye Business Communications. The loudspeakers were

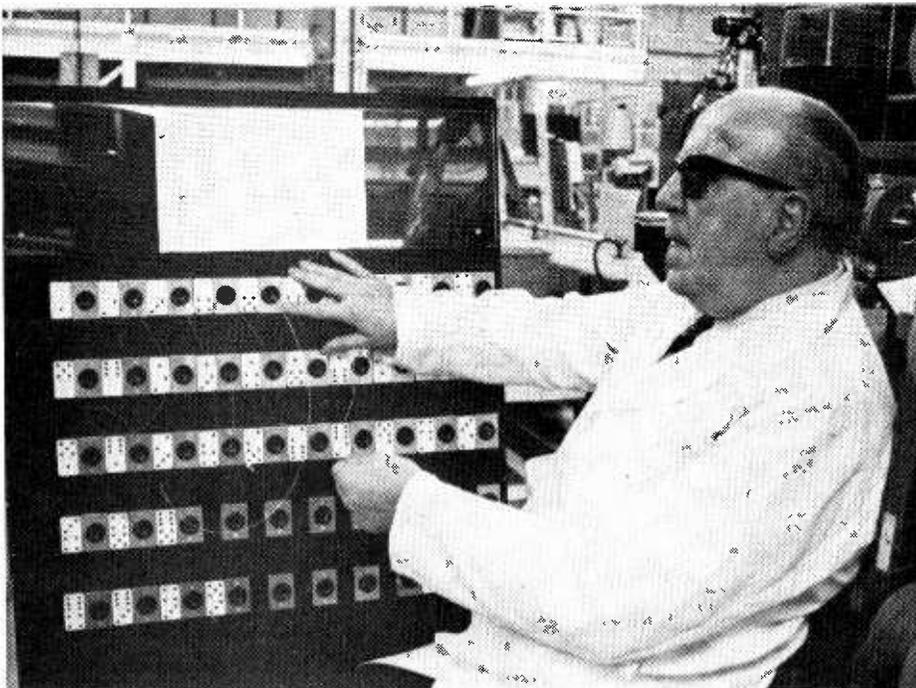
designed for the Abbey by Mr Paul Taylor who donated the patent rights to the Dean and Chapter. The loudspeakers were constructed and tested by the Department of the Environment Building Research Station in collaboration with the Institute of Sound and Vibration Research at Southampton University.

Fingerprint file

The Metropolitan Police at New Scotland Yard in London are to be provided with an information system called Videofile at a cost of approximately £2M over the next two years. The system involves video recording techniques and will be used by NSY to assist its fingerprint identification process. The system will store the fingerprint impressions on magnetic tape so that they may be retrieved rapidly for visual comparison. Impressions are presently held in document form. Under the new system, incoming requests for fingerprint identification – both normal ten-finger "rolled" impressions and also "latent" fingerprints left at the scene of a crime – will be classified for search purposes by fingerprint officers. The police national computer will then be used to produce possible matches from ten fingerprint and display the possible matches on a cathode-ray tube screen for visual comparison by experts.

The primary objective of the system is to extend the capability of the NSY staff of highly-trained fingerprint experts in coping with the anticipated file increases, which could be about 3.5 million fingerprint sets within a decade, creating problems of space and qualified fingerprint staff availability if the existing manual system were to be used. Videofile is a trade mark of the Ampex Corporation.

Albert Stroud, a war-blinded employee of M.E.L. Equipment, operates a set of Braille dominoes, each indicating a different wire needed to be cut for radio receivers. This involves handling 48 different colour coded leads in ten different combinations containing from 7 to 100 different wires where lengths vary between 4in and 2ft. Quite a task for a blind man.



Briefly

Royal Television Society Convention. Television and the future needs of the public will be the theme of the third Royal Television Society bi-annual convention at King's College, Cambridge, September 18-21 this year. Mindful of the work of the Annan Committee looking into the future of broadcasting, the Society hopes to create a forum in which the public needs from television can be stated and the response of the broadcasters obtained.

Variable frequency oscillator for the amateur

A phase lock loop design using discrete components

by I. J. Dilworth, B.Sc.,

Department of Electrical Engineering Science,
University of Essex

The need for a flexible, stable and variable frequency source for use in a transmitter or receiver local oscillator in an amateur band station has never been more evident than at the present time. The approach of generating a 70 MHz carrier facilitates operation on 144 and 432MHz by multiplication, and by using a v.h.f. local-oscillator receivers of high performance and single conversion are possible.

A useful way of generating stable high-frequency signals which are frequency agile is to arrange a lower-frequency oscillator to control a higher one such that the latter assumes the stability of the former. By phase comparison it is possible to derive a controlling element which when applied to the high-frequency oscillator, via a feedback loop, performs this function. It is relatively easy to achieve a high order of stability in a v.f.o. at a low frequency but it is not so straightforward when the frequency approaches the megahertz region. It is true that with due care the problem is not so awesome; nevertheless if the frequency is kept below the MHz region repeatability of results is assured, particularly when the constructor has limited time and equipment.

In this design an 800kHz v.f.o. is used to control a 72MHz v.c.o., the resulting

This novel approach provides stability by using a low-frequency control oscillator.

error in stability being only ten times that of the 800kHz oscillator.

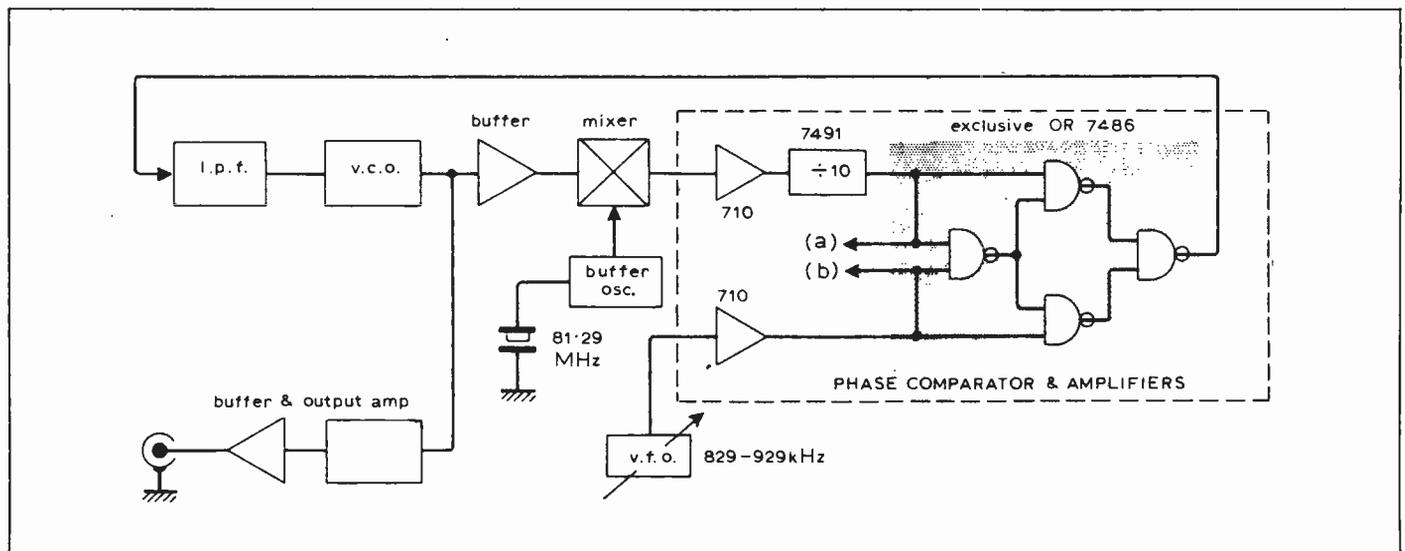
Oscillator

Reference to Fig. 1 will show how locking is achieved. A sample of the 72 MHz signal is mixed down to 9.29MHz with a quartz crystal local oscillator. This signal is then amplified and squared through a digital comparator before being divided down in a decade counter. The resulting 929 kHz signal is then applied to one input of an exclusive OR gate employed as a comparator. The other input is supplied with the v.f.o. signal via the same type of digital comparator to bring it up to the logic levels. The output of the gate is proportional to the phase difference between the two signals and after filtering, to obtain only the d.c. com-

ponent, this voltage is applied to the v.c.o. to maintain an in-phase signal with that of the v.f.o. Any drift in the crystal oscillator is also compensated for in the v.c.o. because it will always try to maintain a zero phase shift with the v.f.o.

The output at 72 MHz is first fed to a buffer and then to a frequency doubler, because in this application it was desired to produce a signal in the 144 MHz band. Since the intermediate frequency is divided by ten in the decade counter, variation of the 800 kHz v.f.o. is effectively multiplied by ten. For a one megahertz covering at 72MHz the 800 kHz v.f.o. needs to tune only over 100 kHz. Clearly drift of the v.f.o. will also be multiplied by ten at this output frequency, but it is relatively easy to construct a stable oscillator at 800 kHz and a suitable design is included. To avoid problems of modulating the 72 MHz signal with either the 63.71 MHz crystal oscillator or the 8.29 to 9.29 MHz i.f. being generated it is necessary to ensure high isolation between the mixer and the output stages. Therefore, one must be very careful with layout and avoid any urges to economise with the buffering. Whether the application be a transmitter v.f.o. or a receiver local oscillator cleanliness in the output spectrum is essential. One could simplify the

Fig. 1. Block diagram illustrating how phase locking is achieved.



buffering circuit but it is not everyone who has a spectrum analyser necessary to set up the unit if the full circuit is not employed. All that is required in setting up the v.f.o. is a grid dip oscillator and a receiver.

Circuit units

The crystal oscillator uses an available unit and is intended for fundamental operation at 63.71 MHz. See Fig. 2. One could use a lower-frequency unit with the appropriate multiplier chain, provided suitable filtering was employed before the mixer, but if one is starting from scratch then it is nearly as cheap to use a 60 MHz crystal rather than a lower-frequency one. The oscillator is straightforward — the tuned collector resonates at the fundamental frequency and the class B buffer stage provides the correct injection for the mixer with the additional filtering of its tuned-circuit.

The mixer consists of a gate injected f.e.t., the tuned circuit in the drain being broadly resonant over 8 to 9 MHz, providing a load for the i.f. signal. The transformer coupled output is then fed into the SN72710 comparator which squares and amplifies the signal prior to the decade counter which consists of a SN7490. The SN7486 gate performs the

comparison of the i.f. signal and the v.f.o. The comparator circuit has proved very useful in practice because the SN72710 devices have a bandwidth up to 30 MHz and are insensitive to input level changes above a threshold of around 10mV. One practical point to note however is that since the outputs of the gates are square waves, screening of the computer output before it reaches the amplifier and low-pass filter is essential otherwise there may be unwanted signals in the station receiver.

To obtain the required swing from the v.c.o. it is necessary to amplify the voltage from the 7400 output. This is conveniently achieved with an operational amplifier whose output is filtered with the network R₁₅-C₁₈ which is the low-pass filter. The variable resistor on the non-inverting input of the 741 provides for initial setting of the output voltage, while R₁₆ provides a high impedance to the tuning diode and the v.c.o. but does not upset the d.c. bias to the tuning diode because there is very

little current needed in this configuration. The diode (D₁) suggested allows sufficient swing in capacitance at 70MHz to provide a 1 MHz change in output frequency with the voltage applied. It is possible to use a 1N916 variety and achieve similar results but with reduced capacitance swing and hence frequency variation. This is because C₁₉ has to be increased in value to allow the Vackar oscillator to oscillate with the lower Q factor of this type of diode. However, if a smaller swing in frequency can be tolerated this approach is possible.

The output of the oscillator is split two ways: one path goes to a source follower whose output is fed to two more buffers before the mixer; the other path goes to the output amplifier chain. Variations in the circuit of the output chain are unlimited and one design is shown in Fig.2. Transistor Tr₉ is a buffer followed by a doubler arrangement providing about 10dBm output power into 50 ohms which is sufficient to drive most class C stages.

Variable frequency oscillator

The oscillator shown in Fig. 3 has proved very stable and trouble free. The

Fig. 2. V.f.o. circuit divided into two parts for mounting on separate boards as shown in Fig. 4 and 5.

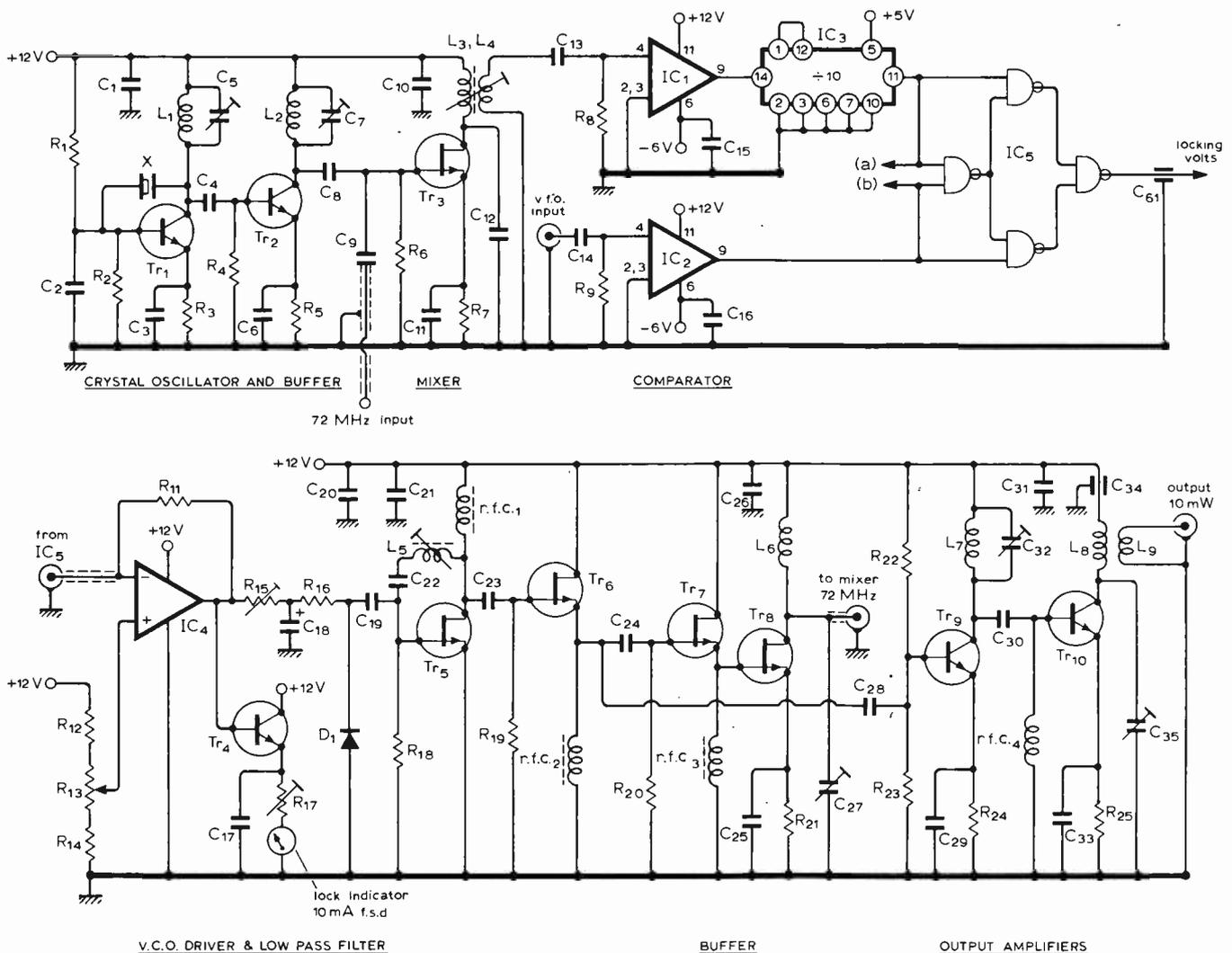


Fig. 3. Stable Vackar type variable low-frequency oscillator tuning between 829 and 929kHz.

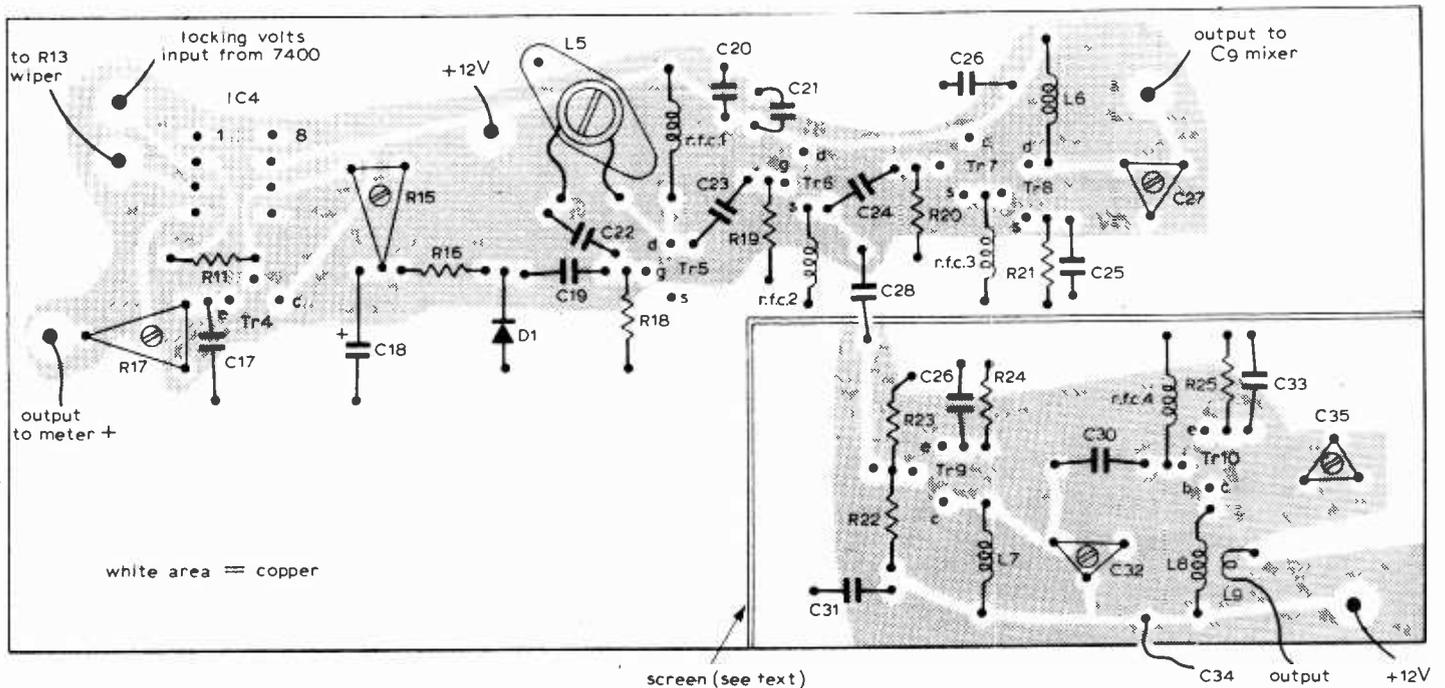
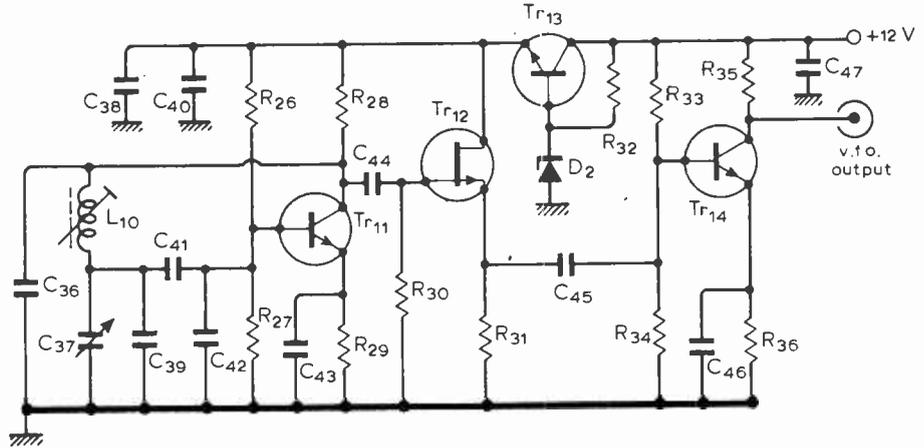


Fig. 4. Printed circuit layout of v.f.o. (viewed from component side) with driver circuit and output amplifiers. The output amplifiers should be screened, as shown, using 1½in high copper strip.

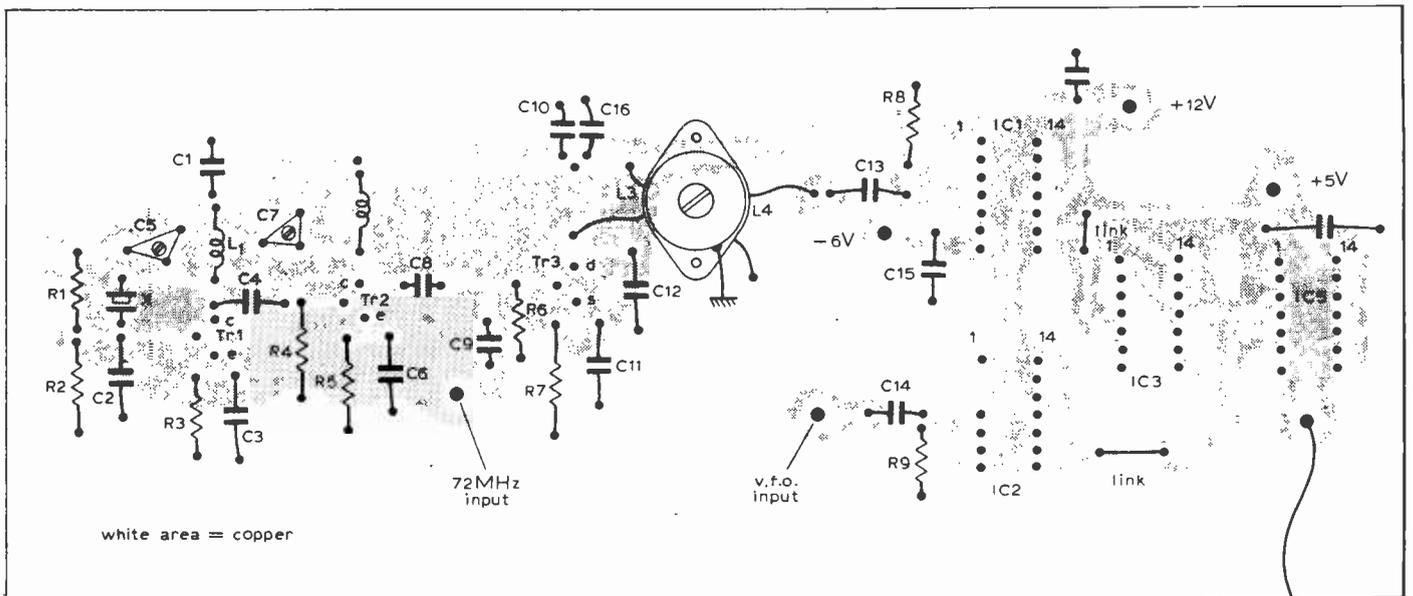


Fig. 5. Printed circuit layout of phase-lock circuitry (viewed from component side) containing crystal oscillator, mixer, and comparator.

the correct size must be employed). Do not forget to join both sides of the board with links in the several indicated places to avoid instability caused by the earth plane floating up at an r.f. potential. Adequate decoupling of the control voltage is essential as any r.f. reaching the v.c.o. at this point will cause wierd and wonderful effects. Because the logic requires five volts it is necessary to use a separate supply for this, not just a zener dropper. It is easy for the supply rail to be modulated with logic pulses, so this must be decoupled efficiently.

The original low-frequency v.f.o. was constructed in a diecast box, measuring two inches deep. This allowed the p.c.b. to be mounted in the centre of the box while the rest of the volume was filled with foam as suggested to produce a stable chamber.

Alignment and testing

First — ensure that the crystal oscillator is working by resonating the coils with a grid dip oscillator or simply rotate the variable capacitor C_5 until a signal is produced. Peak the output of Tr_2 by monitoring the strength of the signal on a g.d.o. Next with no input on the 741 from the NAND gate, adjust R_{13} until the voltage at the output is roughly 4V positive. Leave R_{15} at about $1k\Omega$ and adjust L_5 until a signal is obtained at roughly 72.5 MHz, the frequency does not have to be accurate. Next adjust the voltage on the op-amp to +4V if it has moved, and readjust L_5 . The signal should be reasonably stable and not microphonic. Disconnect C_{28} from the f.e.t. buffer. Connect the output of Tr_8 to C_9 in the mixer, setting C_{27} half way. After checking that the v.c.o. will swing over at least 1MHz between 0 and 12V on adjusting R_{13} , monitor the voltage at the output pin of IC_1 .

Adjust L_3 until the voltage registers 1.5 to 2.0V, indicating that it is squaring the i.f. signal. Swing the v.c.o. over the required range and ensure that it is still squaring by adjusting L_3 as necessary. If there is difficulty with this, increase L_4 with a few more turns. Slight adjustment of C_{27} , which should be almost set, may improve mixing at the edges of the coverage.

Next check that the v.c.o. is tuning over the correct frequency range — 829-929k Hz and apply it to IC_2 making sure that this is squaring. Setting this to 880kHz (mid-band) and the control voltage on IC_4 to about 6V tune L_5 to produce an output at 72.5 MHz, this should be easily receivable using a small wire placed on the bench. Connecting IC_5 output to the input of IC_4 and adjusting R_{13} will produce a lock condition which can be recognized because the pot is being turned and the meter is not moving. Adjusting C_{37} in the v.f.o. will cause the meter needle to move up and down depending on which direction the frequency is being moved. The loop bandwidth is lowered with R_{15} until a clean signal is produced which has

no amplitude modulation. Too much resistance will produce a long lock-in time and this should be set to around midway for best results, although there is no reason why it should not be set higher. Tuning of the output stages is straight forward — Tr_9 being tuned to 72 MHz and Tr_{10} to 144 MHz.

Check that all is well by tuning plus and minus the carrier 8 to 9MHz and adjust C_{77} and C_{27} for minimum signal if necessary.

Using the v.f.o.

Lock should be obtained on switch-on or as soon as the low-frequency reference oscillator is present at the comparator input. The loop bandwidth of the filter is approximately 1 kHz and therefore the locking time is 1ms. The meter being driven from the operational amplifier output serves to indicate that the v.c.o. is tracking with the 800kHz v.f.o. A more elegant way of doing this is shown in Fig. 6. Although this system is not foolproof it could be used to good effect.

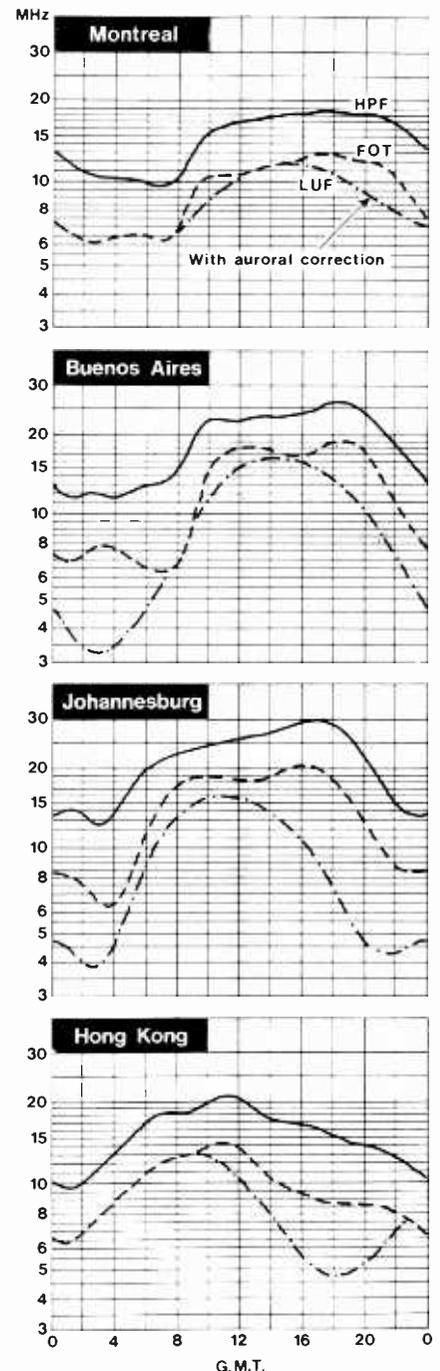
If the v.f.o. is to be used in a transmitter then some means of measuring the output frequency will be needed. If one has a direct reading frequency counter at the output there is no problem. Nevertheless it is possible to use a lower-frequency counter indirectly. For example one method is to measure the 829 kHz v.f.o. frequency and add to this, in the counter, the difference between this and the output frequency. An output frequency of 72 MHz corresponds to a frequency of 829 kHz in the v.f.o., if this is stored in a counter and a frequency of 6.37 MHz is injected, before the counter is reset. The true 72MHz output frequency will result. A possible modification to some frequency counters is shown in Fig. 7.

Depending on the frequency at which it is desired to radiate, buffering and frequency selective stages must be incorporated at the output before it is used with an aerial system. The output stage as it stands produces a component at 144MHz, approximately 18dB stronger than the 72MHz component, and was intended for the first in a chain of class C stages. Frequency modulation is easily applied to the v.c.o. by introducing a few millivolts onto the tuning diode. Care should be taken with screening however if this is tried.

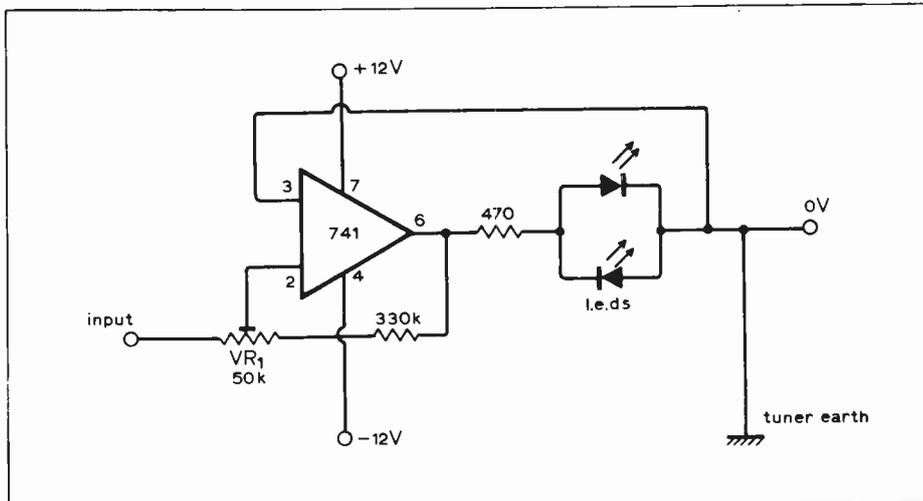
HF predictions

Magnetic activity commented on in recent months has diminished rapidly. September is usually a good propagation month so generally improved conditions should prevail despite it being predicted as the month of sunspot minimum.

Temperate latitude zones will benefit most because magnetic activity has smaller effect at low latitudes and solar activity a greater effect at high and low latitudes.



Circuit Ideas



Simpler f.m. tuning indicator

Several circuits of tuning indicators have been featured in recent issues of *Wireless World*, and much comment has been aroused. The one shown here will no doubt attract its share, but it does have some good points.

It is simple to build, inexpensive and reliable. Operation is also simple. With the l.e.d.s fitted at either end of the tuning scale the cursor is moved away from whichever l.e.d. is on, to a dead spot, i.e. both off, which is the correct tuning point.

Tuning for "lights off" has two main

advantages. Firstly, current consumption is minimized. Both lights can not be on together, and on-tune current is about 2mA. Secondly, a slight drift off-tune, or mis-tuning, is more readily seen. A light coming on, even slightly, is more obvious than one going off or at less than full brightness.

The circuit shown requires a dual polarity power supply, but if a suitable supply is not available in the tuner/amplifier a sub-miniature transformer is adequate.

The circuit including power supply

can be built on a scrap of Veroboard. Component types and values are not critical, nor are power supply voltages. VR_1 is adjusted to give a dead-spot on tune, wide enough to stop the l.e.d.s flickering with loud speech or music. The 470ohm resistor is included to limit the l.e.d. current to a safe value. It may be altered to suit individual tastes in brightness, or where different supply voltages are used.

H. Hodgson,
Thornaby,
Cleveland

Tach-dwell meter

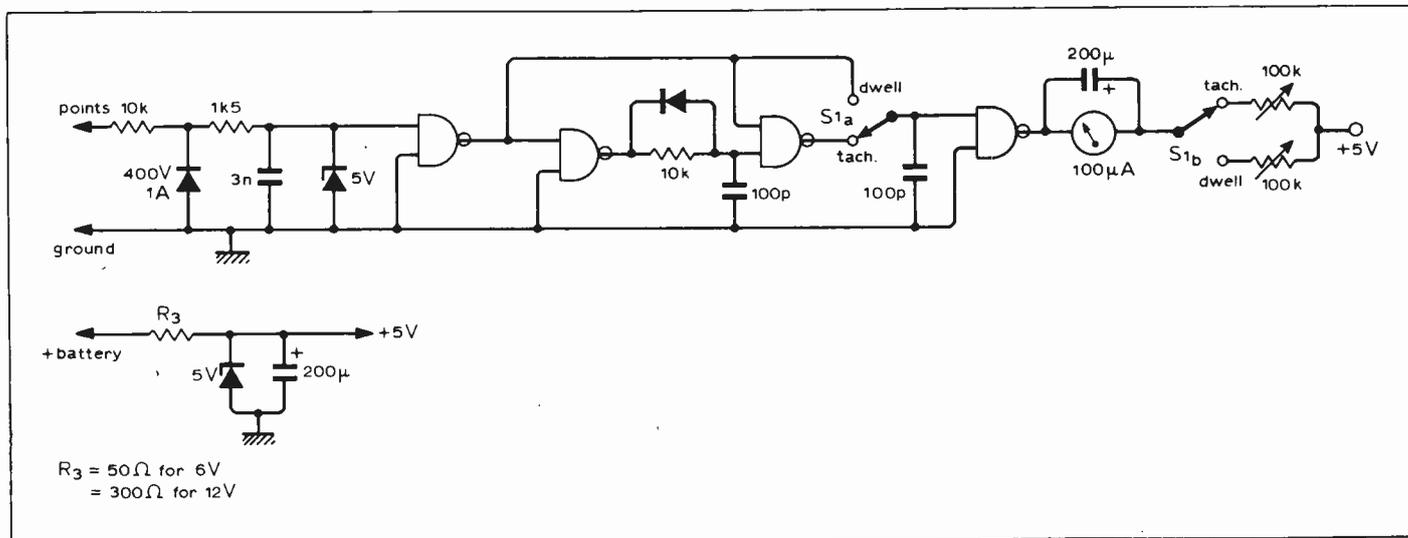
I've been looking for a tach-dwell meter that would use few parts, but would still give me high accuracy and reliability and be relatively inexpensive. I finally built the meter shown which uses a SN7402 NOR gate as the major part. The advantages of this circuit are simplicity, low cost, and high reliability. Reliability comes from the excellent wave shaping and constant amplitude. Also there is no internal battery to wear out. As a point

of interest, the zeners need not be used. The base-emitter junctions of most silicon transistors are 5 volt zeners.

Besides using a known good tach-dwell meter to calibrate the unit, a signal generator can be used. Select the maximum rev/min to be indicated, multiply this number by the number of cylinders, then divide by 120. The answer is the frequency in Hz. A further possibility is use the a.c. mains. Connect

"points" and "ground" across the secondary of a 24-volt transformer and use a battery for power. Select a maximum rev/min of 2,000 and adjust for a reading of 1500 rev/min (for a four cylinder engine; for a six cylinder adjust for 1,000 rev/min and for eight cylinders adjust for 750 rev/min, with proportionately lower f.s.ds).

N. Parron,
Eynsham, Oxfordshire.



Letters to the Editor

ANALOGUE VS DIGITAL READOUT

Your editorial in the July issue reminds me of the problems which arose in the electrical component industry when it was thought that digital readout production test instruments were much quicker than analogue. The female testers soon "told us where to get off" as their throughput and bonus earnings fell considerably because of the time it took to determine a simple (say) resistance measurement of "not greater than" or "not less than" on a digital type ohmmeter, and to put a tolerance on of 10% really confused the issue. Coloured pass-band bands on an analogue indicating meter take all the thinking by operators out of the task, and operators get so skilled in judging the speed the needle moves across the red/green coloured scale that they do not wait for the pointer to stop, knowing at once whether it is a "pass" or "fail". And, woe betide the test gear engineer who uses a poorly damped instrument on a "go/no go" test set.

Digital instruments are best used where absolute accuracy is demanded, but, then, that costs money!

E. J. Williams,
Emsworth,
Hampshire.

ELECTRODYNAMICALLY INDUCED E.M.F.

I see that the old controversy about induced e.m.f.s has surfaced again as part of some interesting correspondence (D. C. E. Todd and N. G. S. Taylor, letters, July issue).

We may be near opening up again the old dichotomy between the "flux cutting" advocates and the "flux linking" school of thought. I have a leaning towards the "linking concept," because, to me, it is closer to the ideas of the vector field and curl. Regarding the aeroplane conundrum, if you glance at

Fig. 1 then when the wings — as a linear conductor — cross the uniform magnetic field lines, the "cutting" theorists claim that an e.m.f. exists between the ends. I am suspicious of this statement because the induced e.m.f. must be completely round a circuit anyway. This is seen by Faraday's Law — in its differential equation form:

$$\text{Curl } \vec{E} = - \frac{\partial \vec{B}}{\partial t}$$

As readers may know, \vec{E} is a vector describing the electric field strength in units of volts per metre. (Curl just says that crawling along an \vec{E} line (a "line integral") we must go right round in small loops at the point in question, i.e. under these conditions, \vec{E} lines can have no start or finish.) \vec{B} is the magnetic flux density; that is, how much flux is passing through unit area at the said point. To get the total flux, we multiply by the area. The differential with respect to time simply yields the "rate of change" requirement for the physical effects to be observed. The minus sign is Lenz's law, which is required for energy conservation requirements. Now all this is just standard electro-magnetic theory which umpteen textbooks contain. (A good one is "Electromagnetic Waves and Radiating Systems" by Jordan, Prentice Hall.) But a closer examination of the equation gives us plenty of clues

to the problem. We only get an e.m.f. round the circuit if we *change* the magnetic flux going *through* the plane of the circuit at right angles. (Vector tyros might note that \vec{B} can only be related to $\text{Curl } \vec{E}$, by a normal or right angles relationship. The operation "Curl" is a *vector product* type.)

Returning now to the magnetically screened return lead for the voltmeter. Does it make any difference to screen it? I suggest no, as a glance at Fig. 2 shows. The presence of the high permeability material will certainly distort the steady magnetic field, but if you imagine the whole "circuit" to move at right angles through the field, then a moment's reflection should convince you that the *total flux linking the circuit is constant*, therefore

$$\oint \frac{\partial \vec{B}}{\partial t}$$

is everywhere 0 and therefore the integrated $\text{Curl } \vec{E}$ is never anything but 0. Even on the "cutting" view, just as many lines will enter the area by crossing the "wing" as will leave the area by moving through the screen and second conductor. From whichever view, no flux linkage change occurs, therefore no e.m.f. is observed.

*Really it is the $\frac{\partial \vec{B}}{\partial t}$ integrated over the area.
K. L. Smith,
University of Kent at Canterbury.

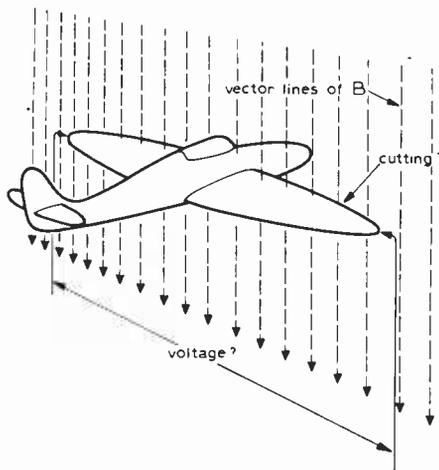


Fig. 1

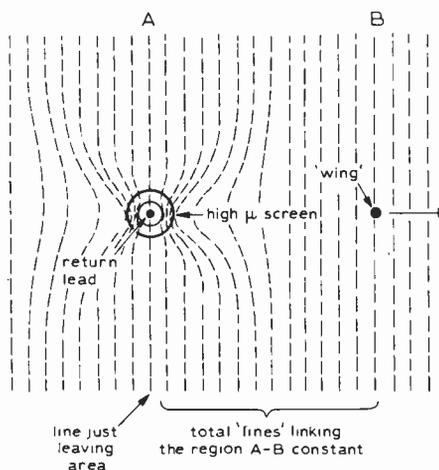


Fig. 2

I was interested to see the recent letters in your journal regarding movement through a magnetic field. The initial statement that no current is induced in a circuit in motion through a constant magnetic field is correct. This rules out the use of moving-coil voltmeters.

The same argument applies equally to the arrangement proposed by Messrs Todd and Taylor since the presence of screening material does not affect the constancy of the magnetic field within the voltmeter circuit.

Possible solutions and a deeper insight into the problem are gained by a proper relativistic treatment of the situation. Because the horizontal magnetic field is relevant, I shall consider a simplified case in which the magnetic field appears to an observer stationary with respect to the earth to be vertical with intensity B . The field seen by an observer on an aeroplane moving horizontally with velocity v is found by performing a Lorentz transformation. The result of the calculation shows that the electromagnetic field has two components: a vertical magnetic field of intensity γB and a horizontal electric field of strength vB/c where $\gamma = (1 - v^2/c^2)^{-1/2}$ and c is the speed of light. The electric field seen from the aeroplane is no mere mathematical illusion but is as real as the earth's magnetic field itself.

Once the question is seen in this light the answer is simple, at least in

principle, since the motion of the aircraft may be measured by any device capable of being affected by an electric field. Such detectors could be made, for example, from semiconductor elements used like field effect transistors or from electron beam deflection systems such as cathode-ray tubes. It is another question, however, whether such a small electric field could be measured in practice.

Colin R. Masson,
Edinburgh

Concerning the query put forward by D. C. E. Todd and N. G. S. Taylor in the July issue (whether the e.m.f. generated by a conductor passing through the earth's magnetic field could be measured if the voltmeter leads were screened by high-permeability material), since the loop formed by the conductor and voltmeter leads would still embrace a constant magnetic flux one suspects the answer would still be no.

There is, however, a different approach to the problem. If a meter were placed at the *centre* of the conductor, the arrangement would be similar to a fixed dipole exposed to the magnetic component of a radio field. But because in this case the field would be both weak and unvarying, and traversed at a speed far below that of radio waves, the problem of measuring the e.m.f. generated might not be readily soluble, unless perhaps the "dipole" was rotated at as high a rate as mechanically reasonable, so converting the "signal" (taken off from slip rings) from d.c. to a.c. Does anyone see any objection in principle?
"Cathode Ray"

TELETEXT DEMONSTRATION

I was rather surprised to read, in your report on the 1975 Spring Trade Shows in the July News of the Month, some very unfavourable comments on the Teletext receiver that Philips were showing.

The receiver shown was certainly an experimental one, but apart from a portion of decorative trim becoming detached, would certainly not merit the description "battered."

We believe that the reason why it was not displaying a Ceefax page at the time your reporter saw it was probably because the "After hours" button had been depressed. This provides synchronisation and enables a page to be retained in the memory and displayed when there are no broadcast transmissions. However, when an attempt is made to call up another page, the result is a blank screen since the "After hours" button has disconnected the broadcast signal.

During the experimental transmissions, decisions obviously have to be made as to what features are to be included and how far the operation of the receiver has to be proof against such effects. Obviously, the Teletext receiver we were demonstrating will have to be modified to avoid this sort of thing, and indeed we have to decide whether "After hours" operation is a worthwhile feature.

Meanwhile, we would like to assure you that great interest was shown by our dealers in this receiver, as appears to be the case wherever Teletext is demonstrated.

M. A. E. Butler,
Philips Electrical Ltd,
Croydon.

DOLBY KIT FILTER ADJUSTMENT

The use of the BBC test transmissions seemed to me to be a little hit-and-miss for setting-up the 19kHz filter of the Dolby noise reducer (July issue) since the vital zero modulation part only lasts for about two minutes. I also did not have a suitable signal generator available.

However, a little thought showed that a precise 19kHz signal was available from pin 10 of my MC1310 stereo decoder when receiving a stereo signal. Since this is the signal that the filter is required to attenuate it seemed logical to use this for alignment purposes. The signal was applied with a 2M Ω potentiometer in series and alignment was easily completed using the signal generator instructions.

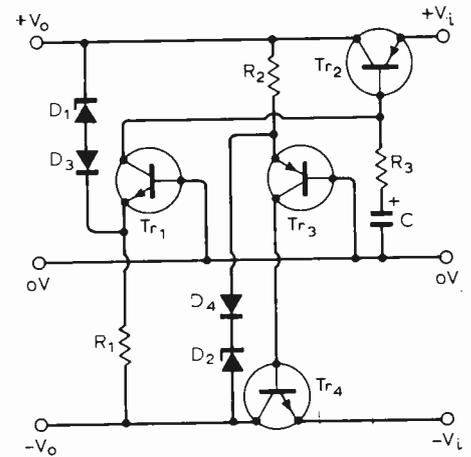
There was possibly some modulation of the signal as the meter flickered slightly, but in spite of this the null was very precise.

Your readers may find this of interest to enable them to set up their kits without having to wait for Radio 3 to close down.

M. S. Maisey,
Coulson,
Surrey.

POWER SUPPLY PROTECTION

The voltage stabilized, symmetrical power supply described by O. Holmskov in the May 1975 issue (Circuit Ideas, p.226) can also be made short circuit proof by the addition of two diodes. If the positive and negative rails of the original circuit are shorted together the zener diodes D₁ and D₂ can be destroyed by the excessive currents flowing along the paths provided by Tr₁, D₁ and D₂,



Tr₂. This can be prevented by the inclusion of diodes D₃ and D₄ as shown in the accompanying diagram. The inclusion of these diodes naturally increases +V₀ and -V₀ slightly due to the diode forward voltage drop, but otherwise the circuit operates as described in the original article. Since carrying out this modification it has not been possible to damage the circuit in any way.

Lothar Bischoff and
David W. Branston,
Erlangen,
W. Germany.

GOOD SERVICE

I have recently purchased test equipment from John Crichton, 558 Kingston Road, London. I should like to place on record the first class after sales service that I have received from Mr F. R. Galka, their manager. I have received a photostat copy, and other components, despite the fact that the equipment is ex-service.

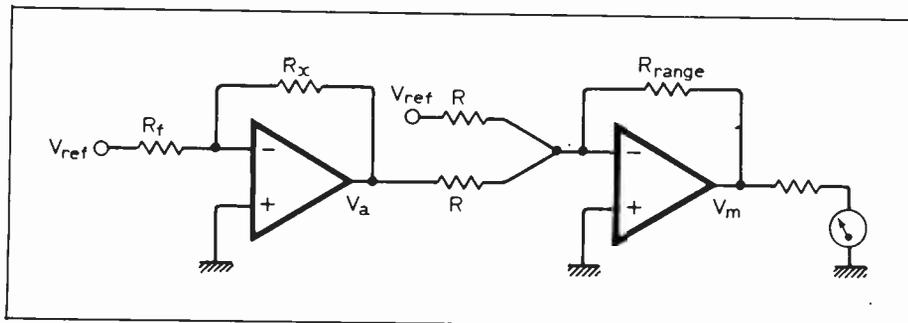
It is very rare these days, as you may well know, to receive such service, and good to know that we still have businessmen like F.R.G.

F. V. Mourant,
St Peter Port,
Guernsey.

RESISTANCE COMPARATOR

The Letters column of W.W. can be relied upon to thump the slightest signs of carelessness or rashness on the part of authors in previous issues. This time I did not even get four weeks' grace, for only a few days after publication of the July issue I found a little note pushed under my office door from one of our research students, Mr P. Choi of the Plasma group, which clearly shows my "crime" to be that of gross inelegance.

His reference (below) to the Company concerns his past role as "employee" in my mythical Curly Wire Co. which goes into battle at Project time in



the undergraduate practical class – fastening upon the possible shortcomings of thermocouples from the dastardly rival Layphlat Wire Co. at that time I recall. Mr Choi writes: “Dear Mr Curly,

It is great to see the Co. at work again as indicated by the Resistance Comparator in the July issue of *W.W.* I was, however, a bit disenchanted with the full circuit on p.333 when Fig. 3 looked so neat. After some restless moments in the bath, it came to my mind that by rearranging V_{ref} we could eliminate much of your circuit: (see above)

Then, $V_m = -R_{range}(V_{ref}/R + V_a/R)$

$$= \frac{-R_{range}}{R} (V_{ref} - V_{ref} R_x/R_f)$$

$$= \frac{R_{range}}{R} \cdot V_{ref} \cdot \Delta$$

where $R_x = R_f(1 + \Delta)$ as before.”

Well done, Mr Choi, this is a very elegant scheme indeed.

D. Griffiths,
Imperial College,
London, SW7.

SUICIDE SOLDERING

Mr Parkins' letter on soldering in the February issue raises some nice safety problems. For a person to receive a fatal shock it is necessary to pass some 70 mA, through the body*, some of this appearing as a potential difference across the heart causing ventricular fibrillation.

Since the mains live is referenced to the earth at the star point the risk of electrocution is obviously increased by strapping the operator to earth. To receive a shock it is only necessary for the person to touch one piece of live equipment, e.g. scope, soldering iron, or convector fire, and the unfortunate operator will be unable to release at least one pole of the circuit. The contact resistance of a deliberate connection to the person will be much lower than, say, a brushing contact.

If it is strictly necessary to earth the operator then the maintenance of the equipment must be of a very high order. I would suggest monthly inspections and tests for earth continuity, earth leakage, and insulation. Alternatively the power supply to the equipment should be isolated from the mains via a suitable transformer and monitored

with earth leakage trips set at 2-5mA. This would render any shocks non-lethal, if somewhat painful. I would further suggest that the habit of sitting on electric heaters should be discontinued and space heating introduced, as the cost of isolating the heaters would be prohibitive.

P. S. Reckless,
Bromley,
Kent.

* Hospital Technical Memorandum No. 8,
H.M.S.O.

DOPPLER DISTORTION

Over the past few months you have printed letters by Mr D. Edgar, Mr J. Moir and “Cathode Ray” prompted by Mr Moir's April 1974 article “Doppler distortion in loudspeakers”. It seems that some of your readers are uneasy about his initial account of this effect, which he attributes to the fact that a low frequency drive signal contributes a component to the cone velocity which Doppler-shifts the frequency of the radiated sound due to a high frequency drive signal simultaneously applied to the cone. In his example (low frequency 100Hz, high frequency 3kHz modulation index $M=0.1$) the frequency deviation swings from +10Hz to -10Hz and back 100 times a second, but the modulated signal consists essentially of just the carrier and two sidebands separated from it by ± 100 Hz. Thus a bandpass filter with a bandwidth of 1Hz and a central pass frequency of 3005Hz would hardly respond at all to the radiated sound, whereas one tuned to 3100Hz would respond strongly. Evidently, as Gabor's acoustical uncertainty relation¹ should lead one to expect, the idea of instantaneous frequency is quite misleading when $M \ll 1$.

An alternative and more basic description of the modulation process avoids these difficulties in interpretation. Suppose that a loudspeaker cone is subjected simultaneously to a complex high frequency signal producing a component of cone displacement $d(t)$, and to a complex low frequency signal producing a component of displacement $d_m(t)$. We assume that all the frequencies involved in $d(t)$ are much higher than all those involved in $d_m(t)$, and that the wavelengths associated with the high frequency signal in air are smaller than the cone diameter. In the

presence of the displacement due to the low frequency signal the radiated high frequency sound waves are given a handicap start of $d_m(t)$, corresponding to a time handicap of $d_m(t)/c$, where c is the velocity of sound in air. Thus the radiated high frequency signal is proportional to $d(t + d_m(t)/c)$. The wave profile is almost unaffected, but there are smooth local fluctuations in the time scale. Reverting to an example equivalent to that given by Mr Moir, if $d(t) = a \sin 2\pi ft$ and $d_m(t) = a_m \sin(2\pi f_m t + \epsilon_0)$, then the radiated high frequency signal near the cone surface will be

$$a \sin 2\pi \left[ft + \frac{a_m}{c} \sin(2\pi f_m t + \epsilon_0) \right],$$

where a_m/c corresponds to the modulation index M . For $M=0.1$ we have a signal of constant amplitude, and a phase which oscillates with the period of the modulating signal, but by a mere $\pm \pi/s$ radians. It is not surprising that many people cannot believe that the ear could detect such subtly camouflaged distortions.

However Mr Moir claims that the members of his listening panel were able to distinguish between music reproduced with relatively high and relatively low levels of “Doppler distortion”, and that they characterised the former as “rough” – surely a surprising comment in view of the smooth and coherent changes in the radiated waveforms produced by the modulating process. It is therefore worth asking whether the original modulation may be made audible by some effect of room acoustics. Has he considered that in a listening room the direct wave and the waves reflected from the walls will in the presence of “Doppler distortion” produce multipath interference effects at the listener's ear, similar to those sometimes encountered in f.m. radio reception?

I have carried out a rough calculation using a single reflected wave, and it appears that, under room conditions, two-path interference can give rise to amplitude modulation sidebands up to a third as strong as the original Doppler modulation sidebands for certain ranges of values of the phase differences between direct and reflected waves of the modulating frequency. The erratic dependence of the amplitude modulation so produced on the modulating frequencies and on the listener's position within the room might well make music reproduced by a speaker which introduces large amounts of Doppler modulation sound “rough”. However, from this standpoint speakers which were otherwise comparable, but produced widely different amounts of Doppler modulation, should be hard to tell apart in an anechoic room.

C. F. Coleman,
Wantage,
Oxon,

Reference

1, D. Gabor, *J.I.E.E.* vol. 93, p.429, 1946.

Transmitter power amplifier design — 1

Circuit techniques and practical considerations for mobile radio h.f. and v.h.f. communications

by W. P. O'Reilly, M.Sc., M.I.E.E.

The Plessey Company Ltd

Power amplifiers for mobile radio transmitters may be classified according to the frequency band covered and to the type of modulation employed. In this series of articles power amplifiers for the high frequency and very high frequency bands are described. The h.f. band extends from 1.5MHz to 30MHz and the v.h.f. band covers frequencies from the top of the h.f. band to approximately 300MHz. These bands are further sub-divided according to the type of traffic they are primarily allocated for; e.g. ship-to-shore radio telephone, mobile radio, broadcast, aircraft bands and several others. The design of a transmitter power amplifier varies considerably with the type of modulation to be employed. In pulsed systems, for example pulse code modulation, the information is contained in the presence or absence of signal and not in the amplitude or phase of the signal. Provided its bandwidth is adequate to permit the required data rate the power amplifier used in such systems cannot distort the signal and so linearity of input/output transfer function is of no consequence. Where several frequencies exist simultaneously in the transmitted signal, as in amplitude modulation (a.m.) systems, it is important that the amplifier should not excessively distort the signal. This restriction is necessary not only to avoid loss of quality or intelligibility of the received signal but also to avoid the generation of new and unwanted signals of sufficient magnitude to cause

In this three part series of articles the author has set out to illustrate the latest design procedures, components and construction techniques used in today's advanced mobile radio transmitters. Part I discusses classification according to frequency bands and the modulation process to be used. The capabilities of present state of the art devices for various applications are indicated. Also in this first article, design procedures are detailed for power amplifiers suitable for operation at h.f. using single sideband modulation. In the second article design procedures for various types of v.h.f. power amplifiers are detailed. Part III details design procedures for some of the special components used in the amplifiers described in parts I and II.

interference in adjacent channels. This type of interference, termed intermodulation distortion (i.m.d.), is discussed in more detail under the heading of power amplifiers for single sideband (s.s.b.) transmitters.

In the v.h.f. bands s.s.b. is not yet in common use but some data modulation systems require high linearity transmitters to avoid distortion which could otherwise result in errors in the received data and spreading of the spectrum into adjacent channels by intermodulation. Many v.h.f. mobile radio links use frequency modulation (f.m.). Instantaneously in an f.m. signal only one

frequency is present and since intermodulation distortion can only occur when two or more signal frequencies are present a linear power amplifier is not required.

Hence power amplifiers may be classified into four groups:— pulse transmitters where amplitude and phase linearity are of no importance; double sideband transmitters in which input/output amplitude transfer function non-linearity (often expressed as a percentage envelope distortion) must be contained within prescribed limits; s.s.b. and certain types of data transmitters in which a high degree of linearity of the amplitude and phase components of the input/output transfer function is required; and f.m. transmitters in which linearity is of no importance. These differ from the first type in that the output power is present at all times during transmission and the amplifier must be adequately rated for continuous operation.

Device capability

The latest generation of mobile radio equipment is almost exclusively solid state in design. Hybrid and integrated circuits are used in the lower power sections of the transceivers and r.f. power transistors form the basis of design of the transmitter power amplifiers. The techniques required to manufacture these r.f. power devices are extremely exacting and are costly to acquire; it is significant that less than half of the world's major semiconductor manufacturers have chosen to enter this sphere of activity.

The design of a power transistor for r.f. operation is a complex consideration of "trade-offs" between such parameters as cut-off frequency, power gain, bandwidth, ruggedness, efficiency and output power. Not all of these may be maximized simultaneously, and the transistor designer has to select his starting semiconductor material, the device geometry and diffusion profiles to achieve a compromise which is suitable to the intended application. Very fine geometry is necessary for a high cut-off

Table 1

Operating Conditions	h.f. 1.5-30 MHz	v.h.f.		u.h.f. 225-400 MHz 15	Microwaves				
		Low band 30-76 MHz	High band 100-175 MHz		1GHz	2GHz	3GHz	4GHz	
Mode Linear (p.e.p)*	Bias class								
	Class A		30	35	15	2.5	1.0		
	Class AB	300	50	70					
c.w./f.m.	Class C	300	80	100	100	40	20	12	6
P _{carrier} a.m.i. p.e.p.	Class C	70	40	30	20				
		280	160	120	80				
pulsed	Class C				300	150	35	20	

*p.e.p. — peak envelope power

frequency; yet fine structures, as well as being the lowest yield types, are generously less rugged than coarser ones. Similarly maximum output power does not coincide with maximum gain or efficiency. The bandwidth achievable is only partly determined by the design of the active device. The transistor package, as well as being the main factor determining the power dissipation capability of the device, has a major effect on the input bandwidth. The inductance of the metalization and bonding wires determines the Q factor of the base circuit which is normally the ultimate bandwidth limitation.

For mobile radio, devices are available for the three usual, most popular supply voltage sources. These are eight volts for personal radio telephones and 12 volts and 28 volts for vehicle equipment. For very high output from a single transistor there are advantages in using a higher supply rail, and some 50 volt devices are becoming available. These are at present finding application in fixed station equipment. Table 1 details the maximum power output capability of r.f. power transistors from h.f. to microwave frequencies for each of the four types of application previously discussed.

Biasing r.f. power transistors

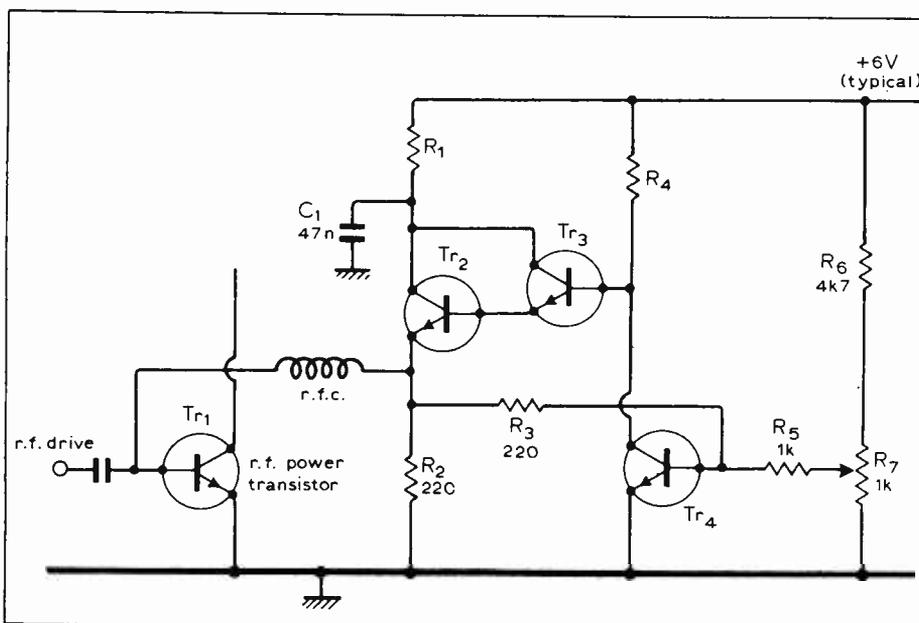
The bias point of a transistor in an amplifying circuit is described in terms of the quiescent collector current, i.e. the current existing before the drive signal is applied. The three main classes of bias used in r.f. amplifiers are: (i) class A in which the base-emitter junction is forward biased so that a large quiescent collector current is obtained. The drive signal modulates this current equally in either sense. Provided the maximum drive signal is not excessive this class of bias provides very linear amplification. The output

Table 2

Operating Condition	Output Power (watts p.e.p.)	Efficiency (%)	two-tone i.m.d. (dB)
Class A	7	20	-45
Class AB Lightly driven	12	40	-38
Class AB Fully driven	25	65	-28
Class C	35	75	-12

power is limited by the steady-state heat dissipation and low efficiency (typically 25% for intermodulation products of level 50dB below full output at h.f.); (ii) class B or class AB in which the base-emitter junction is biased to the verge of conduction — or in the case of class AB to produce a small quiescent collector current — so that as soon as a drive signal is applied the transistor conducts collector current ideally of magnitude proportional to drive signal. A high degree of linearity can be obtained from class B or class AB amplifiers if the bias point is accurately maintained. The efficiency achievable is much greater than for class A bias and hence higher output power can be obtained from the same device; (iii) class C bias is used in f.m. and pulsed continuous wave amplifiers. The base-emitter junction is biased beyond cut-off (generally to zero volts) and the device conducts when the drive signal is increased sufficiently to turn on the base-emitter junction. At low frequencies the collector current is a succession of almost rectangular pulses occurring for short periods centred at the peak of the drive voltage waveform. The collector efficiency can be as high as 80% and is mainly limited by the saturation voltage of the transistor. As

Fig. 1. Temperature compensated bias circuit.



the cut-off frequency is approached, however, the collector current waveform becomes more sinusoidal due to the limited gain of the transistor at harmonics of the drive frequency. The conduction angle therefore increases and may even exceed 180°, and under these conditions the efficiency and output power are less than can be obtained at lower frequencies. Since the drive signal must exceed a certain value before any output power is obtained it follows that the input/output transfer characteristic is extremely non-linear.

Power amplifiers for s.s.b.

Table 2 shows the typical performance of a 2N5707 r.f. power transistor operating at 30MHz. The 2N5707 has a cut-off frequency of typically 150MHz and is characterized for class AB operation up to 30MHz and class C operation above 30MHz. The linearity of an amplifier is commonly tested by driving the amplifier with two equal amplitude r.f. signals separated in frequency by an audio frequency. Spectrum analysis of the output waveform displays the amplifier input signals and also the intermodulation products generated in the non-linearities of the transfer characteristic. If the two input frequencies are f_1 and f_2 it can be shown that third order products occur at frequencies $2f_1 - f_2$ and $2f_2 - f_1$. Similarly fifth order products are $3f_1 - 2f_2$ and $3f_2 - 2f_1$ and so on. These odd-order products are serious since they occur close to the carrier frequency and are not removed by any output filters or tuned circuits, and so they constitute interference to users of adjacent channels. The C.C.I.R. linearity requirements for s.s.b. transmitters are for intermodulation products to be at least -25dB with respect to the level of either of two equal test tones. From Table 2 it can be seen that adequate linearity can be obtained from class AB operation of a transistor specially designed for linear operation, and most mobile radio s.s.b. transmitters use this class of bias in the output stage.

In order that the driver stage distortion shall not significantly degrade the i.m.d. performance of the output stage the linearity of the driver stage must be at least 10dB better than that of the output stage at the peak envelope power (p.e.p.) output. Hence driver stages must either use class AB bias and run well below full output, or, at the expense of increased power consumption, class A bias must be used.

Bias circuits

Since the most important factor determining the linearity of a power amplifier is the class of bias employed, it is essential that the bias point should be accurately maintained over the operating temperature range and under drive conditions. The base current of the power transistor consists of a succession of near half sinewaves the d.c.

component of which must be supplied by the bias circuit; otherwise the transistor would generate a self bias tending to cut-off the base-emitter junction. Many quite complicated circuits have been devised to obtain a temperature stabilized low impedance bias voltage. Most of these systems depend upon a temperature sensitive element such as a thermistor, diode or transistor mounted in close thermal contact with the power transistor. Since it is the temperature dependence of a transistor base-emitter junction which requires matching, this function can be performed very well indeed by using another, but of course much smaller and cheaper, transistor. Fig. 1 shows a bias circuit which uses a transistor, Tr_4 , as the temperature sensor. The current density in Tr_4 is arranged, by suitable choice of R_4 , to be similar to the quiescent current density required in the r.f. power transistor. As a result the temperature coefficient of base-emitter voltage is the same for the sensor as for the r.f. power device and a constant quiescent current is obtained over a wide temperature range. The pre-set potentiometer permits adjustments of the quiescent current to the optimum value and with the circuit values quoted adjustment over the range 0.5 volts to 0.8 volts is possible at 25°C which is sufficient to accommodate the expected spreads in turn-on voltage for silicon r.f. power transistors. The circuit is a feedback amplifier set for less than unity gain, and hence the output impedance is very low (typically 0.01Ω).

The optimum bias impedance for best linearity is usually between 0.3Ω and 1Ω depending on the size of the output transistor, and the r.f. choke between the bias supply and the base of the power transistor may be selected to provide this resistance. Since the current drawn from the bias circuit depends upon both the output power (p.e.p.) and the minimum d.c. gain of the power transistor, the value of R_1 should be chosen to limit the available bias current. This provides some degree of protection in the event of excessive drive signal being applied to the amplifier. Based on a minimum d.c. current gain of 10 for the power transistor, suitable current limits are 8mA per watt p.e.p. for amplifiers running from 28 volts, and 18mA per watt p.e.p. for 12 volt systems.

Broadband matching

Transmitter power amplifiers are intended for operation between defined source and load impedances. Either 50Ω or 75Ω are generally specified. The power transistors, however, are low impedance devices so that impedance transforming networks are required at the input and output of each stage. For a single transistor operating in class AB the load impedance to achieve the required output power is determined by the maximum available voltage swing

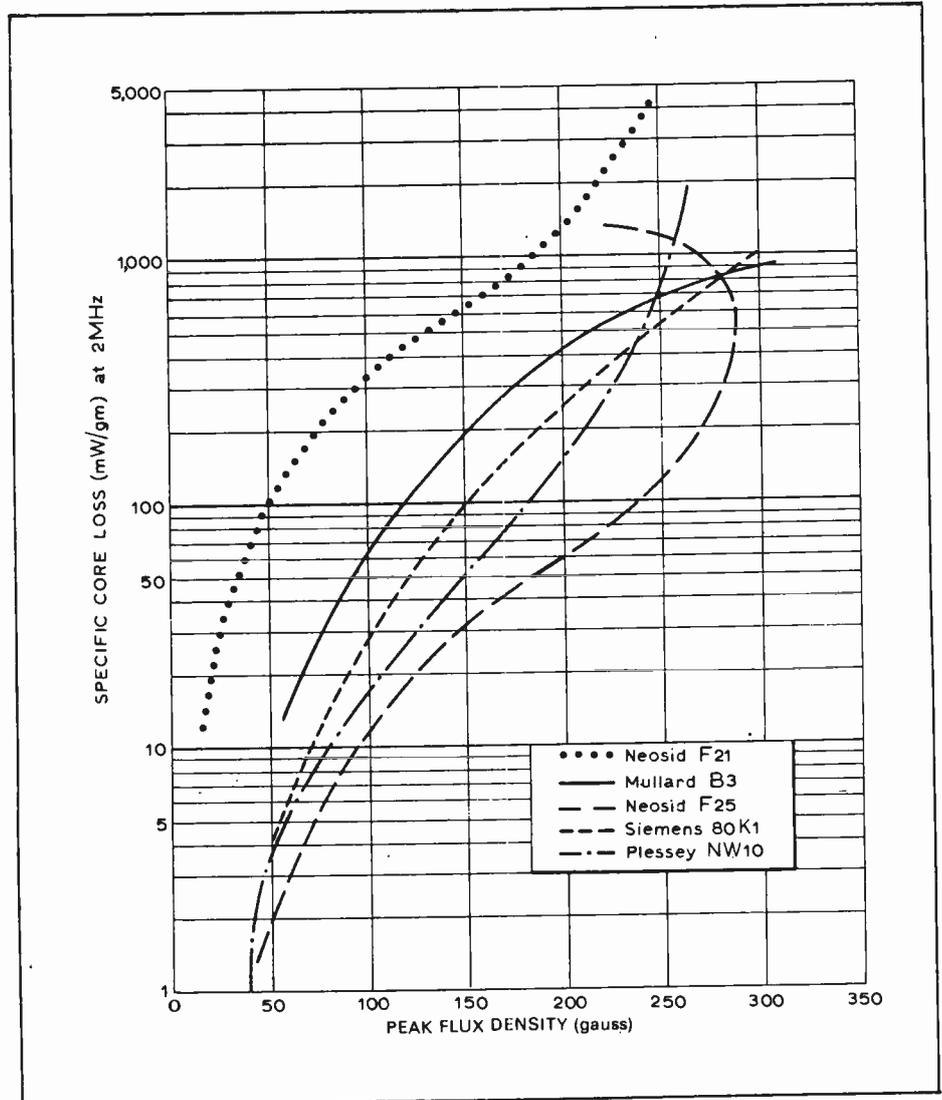


Fig. 2. Ferrite core loss versus flux density for different types of core.

at the collector. For broadband h.f. amplifiers it is normal to operate two transistors in push-pull thus minimizing the second harmonic output. (In a well designed amplifier of this type the second harmonic output is typically 25dB below the fundamental.) The peak available voltage swing is determined by the supply voltage, V_{cc} , and the r.f. saturation voltage, V_{sat} , of the transistors. At present V_{sat} is typically four volts for 28-volt devices used near to their maximum frequency and power capability, and two volts for 12-volt devices. When a push-pull amplifier is operating into a perfectly matched load the output power capability for minimal peak compression (i.e. intermodulation products better than -35dB) is related to the collector load resistance, R_L by the equation:

$$\text{p.e.p.} = \frac{(V_{cc} - V_{sat})^2}{2R_L} \quad (1)$$

If the signal drive is increased the output power will increase by typically 2dB before the i.m.d. products are degraded to -25dB . To ensure that the

amplifier remains linear when running at full power into a moderate load mismatch and also to allow for some losses in the output matching network equation (1) without a correction factor should be used to determine the load resistance. Optimum collector load resistance per transistor at various output power levels is detailed in Table 3 based upon push-pull operation and typical values of V_{sat} . For single ended amplifiers, which are generally used only in narrow band applications, the power output for each stated load resistance is approximately halved.

The required impedance ratio for the output matching network may be determined by reference to Table III. Maximum bandwidth is achieved using transmission line transformers in which the inter-winding capacitance and leakage inductance of conventionally wound transformers are avoided. This is achieved by using transmission lines as the conductors forming the windings. The geometry of this is arranged to provide a characteristic impedance which is optimum for the values of source and load resistance in the circuit. At h.f. ferrite cores are generally used to obtain the high primary inductance necessary for operation at the low frequency end of the band. Toroids, pot-cores or multi-aperture cores may

be used since the high frequency performance is primarily determined by the transmission line windings and not by the ferrite material. Care must be taken in the selection of ferrite cores to avoid operating at too high a flux density as this results in excessive power loss and possible overheating of the windings or even loss of permeability if the temperature exceeds the Curie point. The i.m.d. contribution due to non-linearity in the transformer cores should not be ignored. Fig. 2 shows curves of specific power loss (in milliwatts per gram of ferrite) against flux density for various grades of ferrite measured at 2MHz. It can be seen that when the flux density exceeds about 0.01 weber meter⁻² most ferrites become non-linear and unless special cooling arrangements are made overheating is likely to occur. Of special importance is the need to avoid materials called Perminvar ferrites which contain cobalt. These exhibit extremely low loss in small signal circuits for which they are recommended, but they may be permanently damaged by high flux densities.

The design procedure for broadband transmission-line transformers is best illustrated by an example. The single stage push-pull amplifier of Fig. 3 is required to operate between 50Ω source and load impedances and to produce 50 watts p.e.p. from a 28-volt supply over the frequency band 2MHz to 30MHz. From Table 3 each transistor requires a collector load resistance of 6Ω. The output power from each transistor is combined in the balance-to-unbalance hybrid transformer, T₃. A 4:1 transformer, T₄, then raises the impedance level to 50Ω to match the load. The primary reactance of each transformer winding is required to be at least three times the load resistance presented to that winding at the lowest operating frequency. Thus for each winding of T₃ and T₄ the required primary inductance is 3μH. Selecting initially a 38mm Mullard toroid in grade B3 ferrite (type

Table 3

Output (watts p.e.p.ifiers)	Optimum load resistance(Ω)	
	28V amplifiers	12V amplifiers
10	30	5
20	15	2.5
50	6	1.0
100	3	0.5

FX3027) the required number of turns per winding may be calculated from

$$N = \sqrt{\frac{L}{\mu_o \mu_r} \cdot \frac{l}{A}} \quad (2)$$

where L = required inductance, μ_r = relative permeability = 100 for grade B3, and μ_o = free space permeability = 4π × 10⁷

$$\frac{l}{A} = \frac{\text{magnetic circumference}}{\text{C.S.A. of toroid}} = 2440\text{m}^{-1} \text{ for FX3027}$$

From equation (2), T₃ and T₄ require eight turns per winding. To determine whether a sufficiently large toroid has been selected the maximum flux density, B, must now be calculated from

$$B = \frac{V}{4.4f.N.A.} \quad (3)$$

where V is the maximum r.m.s. voltage across the winding at the lowest operating frequency, f. Using equation (3) a maximum flux density of 0.006 Wbm⁻² is predicted for T₃ and T₄ and at this flux density the specific core loss for B₃ ferrite is 16mW per gm of core material. The total loss in each 18gm transformer core is thus estimated at 290mW, which is an acceptable amount

Fig. 3. Fifty-watt p.e.p. push-pull amplifier.

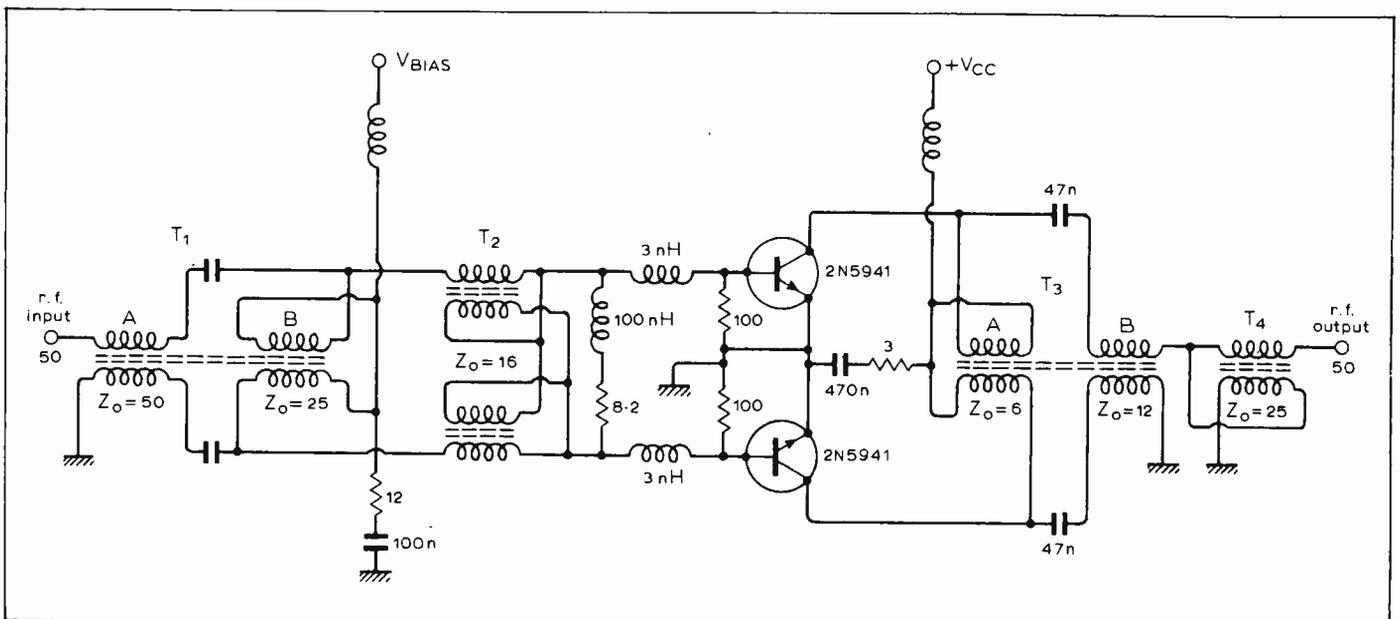
from thermal considerations and will not significantly degrade the i.m.d. performance of the amplifier.

The optimum characteristic impedance for the transformer windings can be shown to be

$$Z_o = \sqrt{Z_{\text{source}} \times Z_{\text{load}}} \quad (4)$$

Transformer T₃ thus requires an impedance of 6Ω for winding A and 12Ω for winding B, while for T₄ a characteristic impedance of 25Ω is required. The transmission-lines may be realised using several high impedance lines (such as bifilar enamelled copper-wires which have typically an 80Ω impedance) connected in parallel. Alternatively coaxial cables or, especially where a low impedance is necessary, copper tapes printed onto either side of an insulating material such as Mylar or glass reinforced p.t.f.e. may be used.

The transformers for the input network, T₁ and T₂, are designed in a similar manner but, since they are operated at a lower power level, smaller ferrite cores may be used. The power from the source is applied to the phase-splitting transmission-line hybrid transformer. Transformer T₁ which converts from 50Ω unbalanced to 50Ω balanced. A 9:1 transformer, T₂, is used to match the source to the transistors which have a typical input impedance at 50MHz of 2.5-j0.2Ω. A very close match is obtained at the top of the frequency band, but at lower frequencies the input impedance of the transistor is higher due to the increased current gain which magnifies the contribution to input impedance made by the emitter ballasting (current sharing) resistors. In order that a reasonably constant input impedance over the whole frequency band is obtained a compensating network is generally required. It is normal to incorporate in this network components to reduce the effects of the gain/frequency slope of the transistor which is usually 5 or 6dB per octave. A



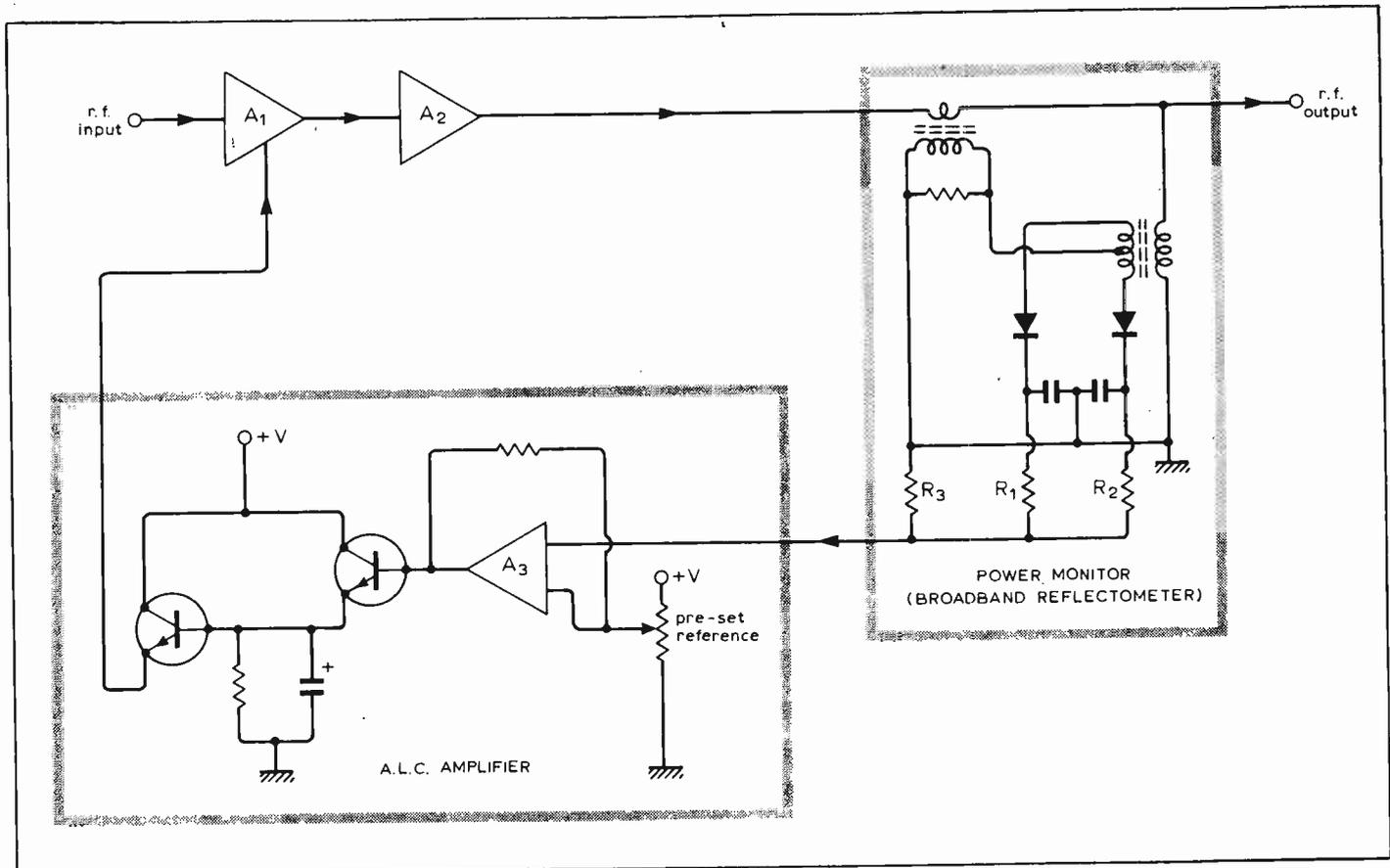


Fig. 4. Typical automatic level control system. A_1 – gain controlled pre-amplifier, A_2 – driver and power amplifier stages, A_3 – operational amplifier.

simple network to provide a degree of compensation has been incorporated in Fig. 3.

Level control and load mismatch protection

The requirement for high linearity in s.s.b. transmitter power amplifiers makes it essential that the output power is maintained within the linear capability of the amplifier, and so some means of controlling the drive power is necessary. To achieve this an automatic level control (a.l.c.) loop is generally incorporated. Most a.l.c. systems consist of three basic units:— a gain controlled element in the forward path of the amplifier, an output power monitor – which can either be a simple output voltage and current detector or, ideally, forward and reflected power detectors – and a comparator or high gain amplifier having an adjustable reference for setting up the output power. A typical system is illustrated in Fig. 4.

The gain controlled element is usually the first block in the amplifier chain. The essential characteristics are rapid response time, linearity maintained over a dynamic range of at least 10dB and a low noise figure. The signal handling capability should be as great as possible since this determines the amount of broadband amplification required and hence, together with the noise figure, ultimately limits the output signal-to-noise ratio. Even when an a.l.c. loop is incorporated it is good practice to ensure that the gain/frequency response of the power amplifier

is as level as possible in order to minimise the radiation of broadband noise power at frequencies remote from the operating channel. A popular method of achieving a linear gain controlled function involves the use of junction diodes in a bridge network. The impedance of the diodes is altered by varying their d.c. bias thus altering the attenuation in the network and controlling the level of input signal to the power amplifier. Very high linearity can be achieved using p-i-n diodes which are now available with carrier lifetimes which make them suitable for operation as variable attenuator elements at frequencies down to 2MHz. Active circuits are available which may perform the variable gain function. The Plessey SL610C integrated circuit r.f. amplifier has a gain control range in excess of 46dB and a noise figure of less than 5dB. For applications in which the broadband gain must be minimized, for example where electromagnetic compatibility considerations are stringent as in equipments for co-sited deployment, the variable gain stage must be capable of handling large input signals. Unless special circuit techniques are used² active devices are generally unsuitable. For such applications the diode bridge, with several diodes in series for each arm is suitable. Other more elaborate systems have been

designed including saturable reactors in attenuator networks and banks of electronically switched fixed attenuators.

An excellent power monitor system consists of detectors sensing the forward component, P_f , and reflected component, P_r , of the output power using a wideband reflectometer. Such an arrangement is shown in Fig. 4 in which the level control signal is a function of both P_f and P_r so that the amplifier output is adjusted to a safe level for any load mismatch. The net output power is related to load v.s.w.r., S , by:—

$$P_{out} = P_f - P_r = \text{p.e.p.} \cdot \frac{4S}{(K^2 + 1)(S - 1)^2 + 4S}$$

The factor K is determined by the relative sensitivity of the forward and reflected power monitors, and in Fig. 4 if $R_3 \ll R_1$ and R_2 then K is the ratio R_1/R_2 . The maximum safe value of K is dependent upon the breakdown voltage of the transistors and the supply voltage, but to ensure that the amplifier linearity is not significantly degraded for mismatched loads it is normal to set a value of K between 1.7 and 2.6 which provides an output power reduction of between 1.5dB and 3dB for a load mismatch of 2:1 v.s.w.r.

The remaining block in the a.l.c. loop is the comparator or a.l.c. amplifier. In the simple arrangement of Fig. 4 a pre-set reference is applied to the inverting input of an operational amplifier and the signal from the output power monitor is applied to the

non-inverting input. If the output power monitor exceeds the level set by the reference the a.l.c. amplifier provides an output signal which reduces the gain of the controlled element and so reduces the drive to the power amplifier. The circuitry at the output of the a.l.c. amplifier provides a rapid response to reduce the gain but will only allow the gain to increase slowly. This is necessary in order to avoid distortion of the r.f. envelope which would, of course, constitute additional i.m.d. To protect the power transistors from damage due to excessive drive or faulty antenna conditions the "attack time" of the a.l.c. system must be as short as possible; the thermal time constant of many r.f. power transistor chips is less than one millisecond, and at low frequencies compared with the cut-off frequency, f_T , of the device localised hot-spots may occur under high stress conditions the thermal time constants for which are often only a few tens of microseconds. In many practical a.l.c. loops, however, in order to ensure that the loop is stable some compromise is made between attack and decay time requirements. These difficulties may be overcome by using a more complex system³ involving two a.l.c. loops, one having very rapid responses, e.g. $1\mu\text{s}$ attack and 10ms decay, and the second having a 5ms attack and 1s decay. When a transient (such as a broken antenna feed cable) occurs the transistors are immediately protected by the fast loop. After a few milliseconds the slower loop takes over control and the amplifier continues to operate safely providing a linear but reduced output.

Combining for higher power

At present the maximum output power which can be achieved from a push-pull stage using two devices is about 500 watts. It is unlikely that such larger transistors will be available for some time for broadband amplifiers due to the difficulty of matching the very low and reactive input impedances. For the present, in any case, to achieve higher powers several transistors must be used. Little success has been obtained by simply paralleling devices, since the input impedance is reduced and matching difficulties occur. Also it is difficult to ensure that the power load is shared evenly between the devices and it is necessary to run the average transistor well below its maximum capability in order to ensure that the most heavily stressed unit is not destroyed. In parallel connected circuits, it is not uncommon for failure of one transistor to result in excessive stressing of the surviving units which fail successively. Such catastrophic failures may be avoided by using the hybrid combiners shown in Fig. 5. Each device is presented with the correct load impedance and if a unit fails the surviving devices are isolated from the faulty unit and continue to drive matched loads. Under these conditions the hybrid combiner diverts

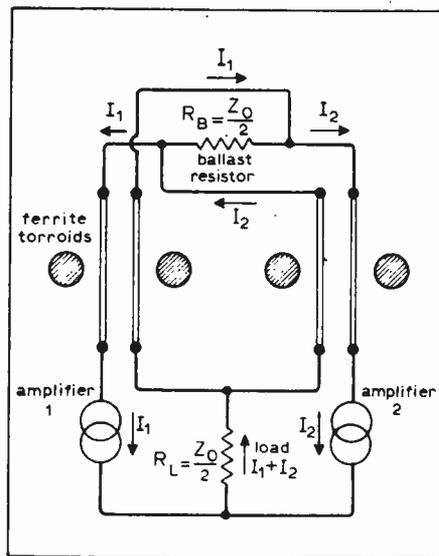


Fig. 5. Zero degree hybrid line combiner.

some of the output power into the ballast resistors and the output of the amplifier is reduced. By combining push-pull modules of approximately 100 watts to 500 p.e.p. capability, transmitter power levels of several kW have been obtained.

The second article in this series will deal with circuit techniques and design procedures for medium and high power solid state amplifiers operating in the v.h.f. bands. A final article will discuss in more detail the design of some of the special components such as transmission-line transformers and strip-line components for v.h.f. circuits which are used in r.f. power amplifiers, together with advice on construction techniques and on precautions to be adopted during initial testing of prototype power amplifiers.

Acknowledgement

The author wishes to express his thanks to The Plessey Co. Ltd. for permission to publish this series of articles.

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1. R.C.A. Application Note No. 3764.
2. Ruthroff, G. L., "Some Transmission Line Transformers," *Proc. IRE*, Aug. 1959, pp. 1337-1342.
3. Matthaei, G. L. Tables of Chebyshev impedance transforming networks of low-pass filter form, *Proc. IEEE*, Vol. 52, Aug. 1964, pp.939-963.

Literature Received

Suppressors for mains-borne interference are the subject of leaflets issued by BHS Electronics (Sales). The devices are balanced twin inductors on ferrite ring cores and a capacitor, epoxy encapsulated in an aluminium can. Working currents are 1A, 3A, 5A and 6A. Available from Bercourt House, York Road, Brentford, Middx. WW401

The range of miniature earphones, headsets, switches etc., made by Danavox are briefly described in Edition 3 of "Electro Acoustic Components Catalogue". The 43-page booklet is obtainable from Danavox (GB) Ltd, "Broadlands," Bagshot Road, Sunninghill, Ascot, Berks SL5 9JW WW402

Collet knobs for panel controls are the subject of the new Radiatron catalogue, publication RE10. A wide range of colours and mechanical styles is described. The catalogue is by Radiatron Components Ltd, 76 Crown Road, Twickenham, Middx. WW403

Prices of Ferranti E-Line transistors (including pro-electron types) are contained in a single price list now issued by Ferranti Ltd, Electronic Components Division, Gem Mill, Chadderton, Manchester WW404

A leaflet on miniature keylock and dual in-line rocker switches is published by Highland Electronics Ltd, 33-41 Dallington Street, London EC1V 0BD. WW405.

A brochure on the Thermalloy range of device and board mounting components and heatsinks is available from Suvicon Ltd, Westminster House, 188-190 Stratford Road, Shirley, Solihull, West Midlands, B90 3AQ WW406

APPLICATIONS

Requirements and setting-up procedure to obtain optimum sub-carrier amplitudes in an f.m. multiplex telemetry system are described in a 12-page application note entitled "Amplitude adjustment of FM Subcarriers", published by EMR Telemetry, Weston Instruments Inc., P.O. Box 3041, Sarasota, Florida 33578, U.S.A. WW407

"Programming Manual for the M6800 Microprocessor" has recently been published by Motorola at £2.50. Available from Motorola distributors.

Methods of achieving savings in industrial electrical energy consumption by the use of power factor capacitors are detailed in the new publication by Bryce Capacitors Ltd, of Helsby, Cheshire WW408

EQUIPMENT

Dc. to d.c. converters for use in telecommunication systems are described in an ITT publication "DC/DC Converters", available from Electrical Product Division, ITT Components Group, Edinburgh Way, Harlow, Essex WW409

Complete details of the range of measuring instruments made by Advance Electronics are given in the new Data Book, which covers oscilloscopes, counters, d.v.m.s, pulse and signal generators and chart recorders. The 58-page 1975/6 Data Book is obtainable from Advance Electronics Ltd, Roebuck Road, Hainault, Essex WW410

A short-form catalogue from the American firm, CVI, offers details of a range of video instrumentation for scientific and industrial use. Instruments described include slow-scan equipment, digitizers, disc memories and X-Y indicators. The catalogue can be obtained from CVI, Inc., P.O. Box 928, Boulder, Colorado 80302, U.S.A. WW411

Electronic circuit calculations simplified

4 — RC combinations in d.c. circuits

by S. W. Amos, B.Sc., M.I.E.E.

Previous articles in this series have dealt with problems involving resistance only and capacitance only. We shall now consider circuits the behaviour of which is determined by a combination of resistance and capacitance. Firstly we shall consider the behaviour of such combinations in d.c. circuits in which valves or transistors are switched on and off by pulses.

Ripple in rectifier circuit. A number of circuits can be reduced to the simple form shown in Fig. 1. This shows a capacitor C which can be charged from a d.c. source via the resistor R₁ (when S is closed) and discharges through the resistor R₂ (when S is opened). One example of such a circuit occurs in a rectifier (Fig. 2) where C is the reservoir capacitor, R₁ represents the forward resistance of the rectifier and R₂ represents the load. The switch S can be

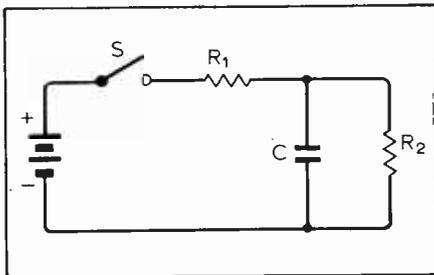


Fig. 1. Simple circuit for charging and discharging a capacitor.

regarded as incorporated in the rectifier which is made conductive and non-conductive by the alternating voltage applied to it.

Suppose we wish to calculate the ripple on the d.c. supply from a half-wave rectifying circuit such as that shown in Fig. 2. For simplicity we can ignore the forward resistance of the rectifier, i.e. in Fig. 1 we can assume R₁ = 0. Thus the charging of C is instantaneous and the whole of the period between successive charges of the capacitor is occupied in supplying power to the load R₂. For 50-Hz mains this period is 1/50th second i.e. 20ms and if the load requires a current of say

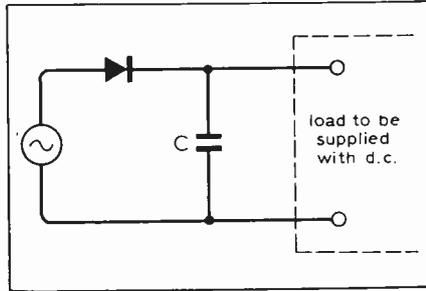


Fig. 2. A half-wave rectifier circuit can be regarded as an example of the circuit of Fig. 1.

50mA the charge removed from the capacitor in this period is given by

$$Q = It$$

$$= 50 \times 10^{-3} \times 20 \times 10^{-3}$$

$$= 10^{-3} \text{ coulomb.}$$

We calculate the drop in voltage across C caused by the removal of this charge from the expression

$$V = \frac{Q}{C}$$

Suppose C is 100μF

$$V = \frac{10^{-3}}{100 \times 10^{-6}} \text{ V}$$

$$= 0.1 \text{ V}$$

Clearly the ripple is inversely proportional to the magnitude of the reservoir capacitor. It is also directly proportional to the load current and to the interval between successive charges. Thus the ripple can be halved by using full-wave rectification for which the interval is only 10ms.

This calculation was made with the aid of two formulae

$$Q = It \text{ and } V = \frac{Q}{C}$$

Many circuits operate by virtue of the charging and discharging of capacitors and the calculation of the values of components to use in such circuits can usually be made using these two expressions.

Time constant. As the next calculation consider the circuit shown in Fig. 3 which can be regarded as part of a pulse amplifier. The input to the base of Tr₁ is assumed to be a pulse signal which holds Tr₁ conductive or cut off. The collector load resistor is R₁, and C represents the total capacitance effectively in parallel with R₁; this includes the output capacitance of Tr₁ and the input capacitance of the circuit Tr₁ feeds. When Tr₁ is cut off, C charges via R₁; when Tr₁ is turned on, C discharges through the transistor.

Consider first the charging of C. C is initially discharged and the full supply voltage V_{cc} appears across R₁. From Ohm's law the charging current is V_{cc}/R₁. When C is fully charged, there is no charging current and therefore no voltage across R₁. Thus we can say that the average current through R₁ during the charging process is V_{cc}/2R₁. How long will this take to charge C to the supply voltage? We know that

$$Q = It \text{ and } V = \frac{Q}{C}$$

Eliminating Q between these, we have

$$t = \frac{VC}{I}$$

Putting I = V/2R₁ gives

$$t = 2R_1 C.$$

This simple relationship is very useful in showing the rate at which voltages can

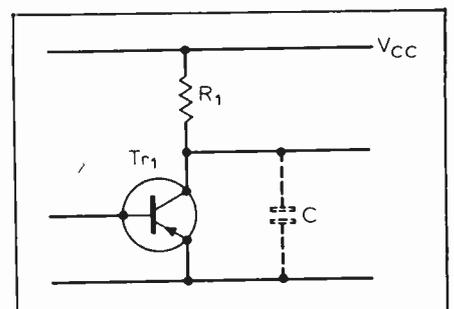


Fig. 3. The stray capacitance C is charged via R₁ when Tr₁ is cut off.

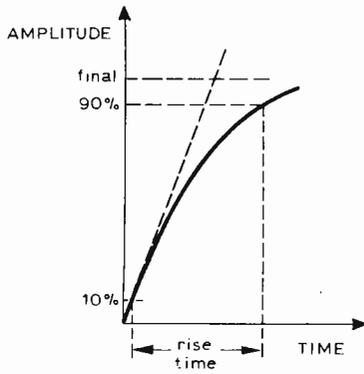


Fig. 4. Definition of rise time of a pulse.

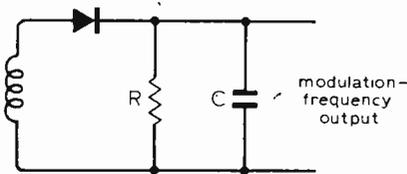


Fig. 5. Essential circuit of a diode a.m. detector.

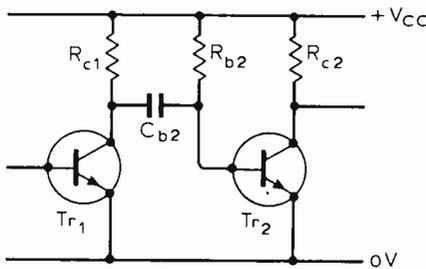


Fig. 6. $R_{b2}C_{b2}$ are the timing components in this simplified monostable circuit.

change across a capacitor. In a practical circuit it is impossible to charge or discharge a capacitor instantaneously (as assumed in the first calculation in this part). It takes time to put charge in or take it out and the time is determined, as this simple expression shows, by the product of the capacitance and the resistance of the external circuit. The product RC is known as the time constant and this expression shows that the time taken for a full charge of the capacitor is twice the time constant.

As a numerical example suppose in Fig. 3 $R_1 = 2$ kilohms and $C = 20$ pF, both typical practical values. C will completely charge in

$$2R_1C = 2 \times 2 \times 10^3 \times 20 \times 10^{-12} \text{ second} = 0.08 \mu\text{s}$$

This is just under $0.1 \mu\text{s}$ – just sufficient, in fact, to be able to reproduce the leading and trailing edges of the line sync pulses in the 625-line television signal.

Rise time. The speed of voltage rises e.g. in the leading edge of a pulse is of great concern in pulse circuitry and is usually measured by the time taken for the voltage to rise from 10% to 90% of the final value. This is known as the rise time. (There is a corresponding definition for the fall time.) In the above derivation we calculated the time taken for the voltage across C to rise from zero to the supply voltage. Our calculation was, however, optimistic because we assumed, by working in terms of the average charging current, that the current falls linearly to zero. In practice the steepness of the current fall becomes progressively less as charging proceeds and it takes longer to charge C than our simple calculation suggests. In fact the rise time is approximately equal to 2.2 times the time constant. This is another relationship which is very useful in calculations on pulse circuits. The rise time for a simple RC combination is illustrated in Fig. 4, which also shows that the waveform of the voltage across the capacitor is not linear but exponential in shape.

As a numerical example suppose a rise time of not less than $1 \mu\text{s}$ is required from a circuit in which the total capacitance cannot be reduced below 25 pF . What is the maximum value of resistance than can be used? From the relationship

$$\begin{aligned} \text{rise time} &= 2.2RC \\ \text{we have } R &= \frac{\text{rise time}}{2.2C} \\ &= \frac{1 \times 10^{-6}}{2.2 \times 25 \times 10^{-12}} \text{ ohms} \\ &= 18 \text{ kilohms} \end{aligned}$$

Constant-current discharge. Let us now consider the discharge of C in Fig. 3. This is achieved by turning Tr_1 on and thus the capacitor discharges, not into a linear resistor, but into the collector-emitter terminals of a conductive transistor. If C discharged into a linear resistor then the current in the resistor would fall as the voltage across it (and the capacitor) falls: the current is, in fact, at all times proportional to the voltage (Ohm's law again!). The transistor does not behave in this way. Although the current through the transistor falls as the voltage across it falls, the current fall is much less than for a linear resistor and, to simplify calculation, it is justifiable to assume that the current through the transistor remains constant during discharge of the capacitor.

Let us assume that the transistor is biased, during discharge of C , to take a current of 10 mA and let us further assume that the supply voltage is 20 . How long does it take to discharge C ? We can use the expression deduced earlier namely

$$t = \frac{VC}{I}$$

Substituting

$$\begin{aligned} t &= \frac{20 \times 20 \times 10^{-12}}{10 \times 10^{-3}} \text{ second} \\ &= 0.04 \mu\text{s} \end{aligned}$$

An important feature of this discharge is that it is achieved by taking a constant current from the capacitor. As a result the voltage across the capacitor falls linearly with time (not exponentially as when the capacitor discharges into a constant resistance). A linear voltage change is, of course, useful in timebase circuits and ramp generators: charging or discharging a capacitor by means of a constant current is the usual way of generating such voltages.

Diode detector circuit. Another commonly-used circuit which operates by virtue of the charging and discharging of a capacitor and in which the choice of the time constant is important is the diode a.m. detector shown in its simplest form in Fig. 5. The problem is to calculate values of R and C suitable for a particular application.

The mode of operation of the circuit has much in common with that of the half-wave rectifier of Fig. 2. The detector input is an alternating signal (the carrier wave) and the diode conducts during positive-going half cycles and charges the capacitor to the peak value of the carrier input. During negative-going half-cycles the capacitor discharges through the load resistor. This is precisely what happens in the mains rectifier circuit but in the detector circuit the input frequency is much higher, e.g. 1 MHz in a medium-wave a.m. receiver.

In the mains rectifier the diode input signal is of constant amplitude: for the diode detector the input amplitude rises and falls about its mean value in accordance with the waveform of the modulating signal. It is essential, to avoid distortion, that the voltage across the capacitor should faithfully follow the changes in the amplitude of the diode input due to modulation. Usually there is little difficulty in following increases in amplitude because the diode forward resistance is low and the capacitor can quickly be charged to a new higher voltage. Difficulties can arise, however, when the amplitude of the diode input falls: to avoid distortion the capacitor must be able to discharge through the load resistance so quickly that it has to be recharged by every positive-going half-cycle even though the amplitude of successive half-cycles is falling. We can estimate the time constant required to achieve such a performance in the following way.

The greatest rate of change of carrier amplitude occurs at the highest modulating frequencies and when these have their greatest depth of modulation. At the highest modulating frequency f_{max} the time taken for the carrier amplitude to change from its maximum to its

minimum amplitude is half the period of the modulating frequency i.e. $1/2f_{max}$. In this time C must be able to discharge completely through R. We have already seen that the time taken for a complete discharge of C is approximately $2RC$. Thus we can say

$$2RC = \frac{1}{2f_{max}}$$

from which

$$RC = \frac{1}{4f_{max}}^*$$

As a numerical example consider the vision detector in a 625-line television receiver. The maximum modulating frequency is 5.5MHz and thus

$$RC = \frac{1}{4 \times 5.5 \times 10^6} \text{ second} \\ = 0.045 \mu\text{s}$$

The product of R and C must not therefore exceed $0.045 \mu\text{s}$. This is a very small time constant and if R is made greater than a certain value C becomes impossibly small. Let us therefore fix C at say 15pF, reasonably greater than the stray capacitance inevitable in the circuit. This gives R as

$$R = \frac{0.045 \times 10^{-6}}{C} \\ = \frac{0.045 \times 10^{-6}}{15 \times 10^{-12}} \text{ ohms} \\ = 3 \text{ kilohms}$$

Now consider a detector to be used in a medium-wave receiver. Due to high-frequency cut-off in the transmissions and in the i.f. circuits of the receiver the highest modulating frequency at the detector input can be taken as 5kHz and, for this value, we have that the detector time constant is given by

$$RC = \frac{1}{4f_{max}} \\ = \frac{1}{4 \times 5 \times 10^3} \text{ second} \\ = 50 \mu\text{s}$$

This is a large time constant which gives considerable freedom in the choice of values for R and C. R should be large compared with the diode forward resistance but, provided this requirement is met, should be as small as possible and a value of 5 kilohms is

commonly used. For this value of R we have

$$C = \frac{\text{time constant}}{R} \\ = \frac{50 \times 10^{-6}}{5 \times 10^3} \text{ F} \\ = 0.01 \mu\text{F}$$

Control of pulse duration. RC circuits play an essential part in the operation of pulse-generating circuits such as multivibrators and the time constants are important because they determine the duration of the generated pulses. This can be illustrated by Fig. 6 which shows an RC circuit $R_{b2}C_{b2}$ coupling Tr_1 collector to Tr_2 base. These two transistors are assumed to be a monostable multivibrator but, for simplicity, the direct coupling between Tr_2 output and Tr_1 input has been omitted.

Because Tr_2 base is connected to its emitter by R_{b2} Tr_2 is normally conductive and the direct coupling to Tr_1 ensures that this is non-conductive. Tr_1 collector voltage is therefore normally at the supply positive voltage V_{cc} and Tr_2 base voltage is normally near the supply negative value: thus C_{b2} is charged to the supply voltage V_{cc} . This is typical of the stable state of the circuit in which it can remain indefinitely. The circuit can, however, be compelled to leave the stable state by a positive-going signal applied to Tr_1 base to turn Tr_1 on. This causes Tr_1 collector voltage to fall and this fall is communicated to Tr_2 base by C_{b2} so cutting Tr_2 off. The effect of this triggering signal is thus to reverse the states of the two transistors. The same effect could also be achieved by applying a negative-going triggering signal to Tr_2 base.

Now consider the circuit conditions for C_{b2} immediately after Tr_1 is made conductive. The terminal connected to Tr_1 collector is now effectively at supply negative voltage: the other terminal is connected via R_{b2} also to supply negative as shown in Fig. 7. C_{b2} is still charged to the supply voltage V_{cc} and immediately begins to discharge through R_{b2} . It is, in fact, the voltage generated across R_{b2} by the discharge current which keeps Tr_2 cut off.

As soon as the discharge is completed Tr_2 begins to conduct again. Thus Tr_2 is held non-conductive during the whole of the discharge of C_{b2} and a positive-going pulse is generated at its collector the duration of which is governed by the time constant $R_{b2}C_{b2}$. The problem is to calculate the values of R_{b2} and C_{b2} required to give a required pulse duration.

We have already deduced a simple expression for the discharge time of an RC combination: it is $2RC$. Thus we can say that the duration of the positive-going pulse generated at Tr_2 collector is given approximately by $2R_{b2}C_{b2}$. As a numerical example suppose these pulses are required to have 10μs

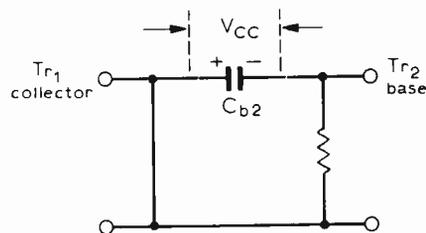


Fig. 7. Circuit conditions for C_{b2} (Fig. 6) immediately after Tr_1 has been turned on.

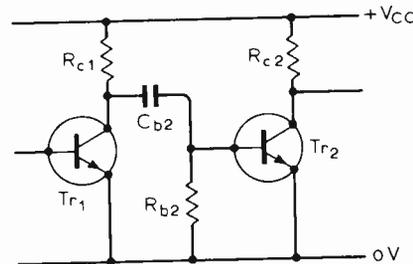


Fig. 8. Variant of the circuit of Fig. 6.

duration. $R_{b2}C_{b2}$ must hence be $5 \mu\text{s}$. If R_{b2} is 5 kilohms, C_{b2} is given by

$$C_{b2} = \frac{\text{time constant}}{R_{b2}} \\ = \frac{5 \times 10^{-6}}{5 \times 10^3} \text{ F} \\ = 1 \text{ nF}$$

Fig. 8 shows a variant of this circuit in which Tr_2 base resistor is returned to supply positive. This does not fundamentally alter the operation of the circuit. Immediately after Tr_1 is turned on the circuit conditions for C_{b2} are as pictured in Fig. 9. If we rearrange this as shown in Fig. 10 we can see that the voltage across C_{b2} and the supply voltage V_{cc} are in series so that the voltage across R_{b2} available to drive discharge current through R_{b2} is double that in the previous circuit and the discharge of C_{b2} is accordingly faster. At the beginning of the discharge there is a voltage of V_{cc} across C_{b2} which keeps Tr_2 turned off. As discharge proceeds the voltage across C_{b2} falls, reaching zero when C_{b2} is discharged. At this instant Tr_2 begins to conduct again and the base current in Tr_2 anchors the voltage at one end of R_{b2} to supply negative. The other end of R_{b2} is still, however, connected to $+V_{cc}$ so that there is still a voltage of V_{cc} across R_{b2} even though C_{b2} is completely discharged.

We can deduce how quickly the capacitor is discharged in this circuit in the following way in which, for simplicity, we will dispense with the suffices to R and C. The voltage across R is $2V_{cc}$ at the beginning of the discharge and V_{cc} at the end of it. The average voltage during the discharge process is thus

* A more rigorous analysis of this problem shows that a more accurate value of the time constant is given by

$$RC = \frac{1}{2\pi f_{max}}$$

$1.5V_{cc}$ and, from Ohm's law, the average discharge current is given by $1.5V_{cc}/R$. If the discharge takes t seconds then the charge removed from the capacitor by this current is given by It , i.e. $1.5V_{cc}t/R$. From the relationship $V = Q/C$ we can deduce that the removal of this charge from the capacitor will produce a fall in voltage across it equal to $1.5V_{cc}t/RC$. Because the capacitor is completely discharged this fall in voltage must be equal to V_{cc} . Thus we have

$$\frac{1.5V_{cc}t}{RC} = V_{cc} \text{ which gives } t = 0.67RC^*$$

Thus the discharge of the capacitor is three times as fast as in the circuit of Fig. 6.

As a numerical example, suppose we wish to generate $50\mu s$ pulses at Tr_2 collector in the circuit of Fig. 8. Then

$$0.67RC = 50\mu s \text{ giving } RC = 75\mu s.$$

A suitable value for R may be 10 kilohms. This gives

$$C = \frac{\text{time constant}}{R} = \frac{75 \times 10^{-6}}{10^4} \text{ F} = 7.5 \text{ nF}$$

Free-running or astable multivibrator.

By combining two circuits of the type shown in Fig. 8 we produce the circuit of Fig. 11, a free-running or astable multivibrator. This generates current pulses continuously in the two transistors and the duration of the pulses can be made any desired value by appropriate choice of the time constants. The duration of the off-period of Tr_2 is determined by the time constant $R_{b2}C_{b2}$ and is given approximately by $0.67 \times R_{b2}C_{b2}$ as shown in the previous paragraphs. Similarly the duration of the off-period of Tr_1 is given by $0.67 R_{b1}C_{b1}$. The sum of these two periods makes up one complete cycle of operation of the multivibrator and thus we have

$$\begin{aligned} \text{free-running period} \\ \text{of multivibrator} \\ = 0.67(R_{b1}C_{b1} + R_{b2}C_{b2}) \end{aligned}$$

from which

$$\begin{aligned} \text{free-running frequency} \\ \text{of multivibrator} \\ = \frac{1}{0.67(R_{b1}C_{b1} + R_{b2}C_{b2})} \end{aligned}$$

If $R_{b1} = R_{b2}$ and $C_{b1} = C_{b2}$ the circuit is symmetrical and generates square waves at each collector. The frequency of the square waves is given by

$$f = \frac{1}{1.33R_b C_b}$$

* The first part of an exponential curve is almost linear and therefore this is quite a good approximation: a rigorous analysis gives $t = 0.69RC$.

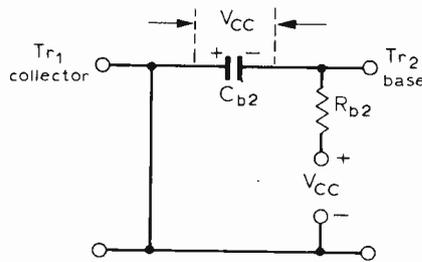


Fig. 9. Circuit conditions for C_{b2} (Fig. 8) immediately after Tr_1 has been turned on.

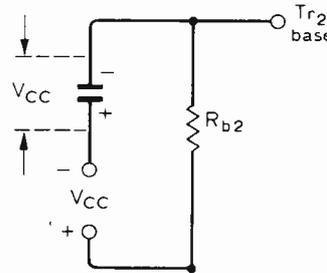


Fig. 10. Rearrangement of Fig. 9 showing the effective doubling of the voltage providing the discharge current.

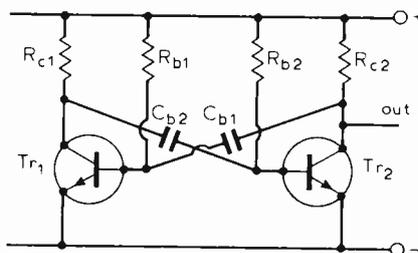


Fig. 11. A free-running multivibrator circuit.

As a numerical example suppose we wish to generate square waves at approximately 50Hz. What values of R_b and C_b should be used? We have

$$\begin{aligned} R_b C_b &= \frac{1}{1.33f} \\ &= 15 \text{ ms} \end{aligned}$$

This is the time constant required to give a free-running frequency of approximately 50Hz. Astable multivibrators are, however, frequently synchronised at the frequency of an externally-applied signal and synchronisation is usually achieved by terminating the unstable periods earlier than would occur naturally, e.g. by applying positive-going sync signals to the base of an n-p-n transistor. Thus the free-running frequency of a multivibrator to be synchronised should be made lower than that of the sync signal and in the above example it would be wise to make the free-running frequency lower than 50Hz by increasing the time constant to say 20ms.

It would appear that any combination

of resistance and capacitance would be suitable provided the product is 20ms. This is not true, however, for there must be a certain relationship between R_c , R_b and C_b to obtain a satisfactory performance. For example when a transistor is cut off its collector voltage rises, causing the base voltage of the other transistor also to rise until this reaches zero (the emitter voltage). The other transistor then begins to conduct and anchors one terminal of C_b effectively at earth potential so that the collector voltage of the first transistor can rise only by virtue of C_b charging through R_c . Thus the collector-voltage rise is slowed down, being governed by the time constant $R_c C_b$. As shown earlier in this article the rise time is given approximately by $2R_c C_b$. For a rapid rise time the time constant $R_c C_b$ must be as small as possible.

There is also a relationship between R_c and R_b . For R_b determines the base current of transistor when it is conducting and a simple application of Ohm's law gives this current as V_{cc}/R_b . The collector current is β times this i.e. $\beta V_{cc}/R_b$. This collector current, in flowing through R_c , must generate a voltage large enough to bring the collector voltage down to zero otherwise the multivibrator cannot operate properly. The collector voltage swing is given by $\beta V_{cc} R_c / R_b$ and this must at least equal V_{cc} i.e. R_b must be less than βR_c . Thus the following three relationships must be satisfied:

$R_b C_b$ is fixed by the required frequency of operation

$R_c C_b$ should be as small as possible

R_b must be less than βR_c

Suppose the supply voltage is 12 and we decide that the collector current in the on transistors shall be 5mA. Then from Ohm's law R_c must be at least 2.4 kilohms. If β is 100 R_b must be less than $100R_c$ i.e. 240 kilohms. A suitable value would be 200 kilohms. We can now calculate C_b thus

$$\begin{aligned} C_b &= \frac{\text{time constant}}{R_b} \\ &= \frac{20 \times 10^{-6}}{200 \times 10^3} \text{ F} \\ &= 0.1 \mu \text{ F} \end{aligned}$$

The rise time is

$$\begin{aligned} 2R_c C_b &= 2 \times 2.4 \times 10^3 \times 0.1 \times 10^{-6} \text{ second} \\ &= 0.5 \text{ ms approximately} \end{aligned}$$

When a transistor is turned on its collector voltage can fall very rapidly because there is no large capacitance to be charged or discharged when this occurs. Thus the voltage pulses generated at the collectors have better fall times than rise times.

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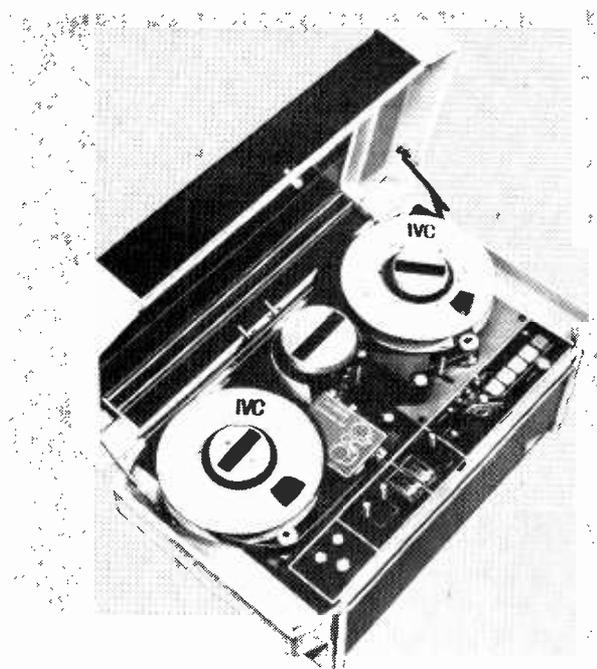
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V.L.F. transmitting aeri-als

An insight for the non-specialist

by R.B.C. Copsey, M.I.E.E.

Merton Technical College, London

Very low frequency waves are used for long range communication where high reliability and freedom from ionospheric disturbances are of prime importance. Carrier frequencies are usually between 14 and 20kHz and because these frequencies are low, and aerial circuits tend to have a high Q, it is necessary to restrict the bandwidth of the transmitted signal, permitting only the use of telegraphy. On/off keying (A1) and frequency shift keying (F1) are both used.

One of the early stations built for shore-to-ship communication was the British Post Office station at Rugby, which first came into service in 1926, operating on 16kHz with the call sign GBR. At the time of its opening it was the most powerful transmitter in the world, employing 54 ten-kilowatt water-cooled valves in parallel³. The station has been in almost continuous service since, although various improvements have been made⁴, notably to its frequency stability, and to its radiated power output which has been increased from 40 to 65kW.

Due to their high degree of phase stability and long range v.l.f. waves are also used for radio navigation systems. The Omega system⁵, will eventually employ only eight transmitters to provide a world-wide navigation system. The eight synchronised transmitters will operate sequentially on a number of carrier frequencies between 10.2 and 13.6kHz. By measuring the phase difference between signals received from pairs of transmitters it is possible to determine lines of position which are hyperbolic in shape, and hence, with the aid of an Omega chart, to find one's position in terms of latitude and longitude. The ultimate accuracy of the system is determined by the phase stability of the transmitters, 1 in 10^{12} , and by the phase stability of the propagation medium. By making suitable corrections for diurnal shifts, it is expected that the overall error in position fixing will not exceed two nautical miles⁵.

Many of the radiation characteristics required for v.l.f. transmitting aeri-als are identical for the Omega system and

for communications. Some of the more important requirements are:

- omnidirectional radiation in the horizontal plane
- vertical polarization
- radiation in the vertical plane should be confined as far as possible to the horizontal component
- high power handling capacity
- high efficiency
- ability to operate continuously in all weathers

Communication systems require a bandwidth of around 100 to 150Hz but changes in carrier frequency are seldom necessary. Omega systems need far less

Very low frequency waves lying in the range of 10 to 30kHz are unique in that given sufficient radiated power, they have a useful range extending continuously to the antipodes.

In effect the D region of the ionosphere and the surface of the earth, most of which is sea water with comparatively high conductivity, form a spherical waveguide, allowing the wave to travel with little attenuation. Gould and Carter¹ show an almost constant average daylight attenuation of about 2.7dB per 1,000km, for ranges from 4,000 to 12,000km from the transmitter at a frequency of 16kHz. This waveguide mode of propagation provides a high degree of phase stability, although there is some phase change at night when the effective height of the D layer rises.

Numerous studies have been made of the behaviour of v.l.f. waves, and changes of phase due to propagation variations can be predicted with reasonable accuracy. One other notable characteristic of v.l.f. waves is their ability to penetrate sea water, enabling submerged submarines to receive radio messages. Attenuation due to sea water varies from about 7dB/m at 40kHz to 3.5dB/m at 10kHz (ref. 2).

bandwidth because each "mark" has a duration of at least 0.9 second; but the aerial system must be re-tunable in less than 0.2 second, as this is the interval between radiation on successive frequencies.

There are two other important requirements for the Omega aerial. One of these is phase stability. If an aerial with a Q of 200, operating at a frequency of 10kHz, is considered, then a 0.1% (10Hz) change in resonant frequency due to changes in temperature, humidity or even the aerial moving in the wind, would produce a phase difference of 20° which is equivalent to moving the aerial 1.67km or approximately 0.9 nautical miles. The phase of the radiated field is therefore controlled by comparing the aerial current with a stable reference and the resulting phase error signal is used in a feedback loop to correct the aerial tuning to maintain a high degree of phase stability. The other important requirement is that the aerial should have only one "phase centre" so that the apparent source of radiation should be a single fixed point, irrespective of the direction of approach.

Aeri-als used in the Omega system are required to radiate about 10kW, which provides a satisfactory signal-to-noise ratio with the very narrow tracking bandwidth, of the order of 0.01Hz, which is used in an Omega receiver. For communication purposes, where receiver bandwidths are of the order of 100Hz for telegraphy, the transmitters need a higher power to achieve an adequate signal-to-noise ratio and powers of 50 to 1000kW are radiated.

The simplest type of aerial which will provide omni-directional radiation in the horizontal plane, with vertical polarization, is the vertical monopole. With wavelengths lying between 10 and 30km, it is clearly impracticable to construct a quarter-wave Marconi aerial, and even with an aerial height, h , of 300m, the electrical length of the vertical conductor, $h360/\lambda$, is only a few degrees.

The radiation pattern in the vertical plane from an electrically short monopole is shown in Fig. 1, where the radiated field strength is proportional to

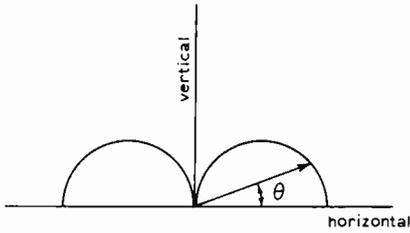


Fig.1. Field strength in vertical plane.

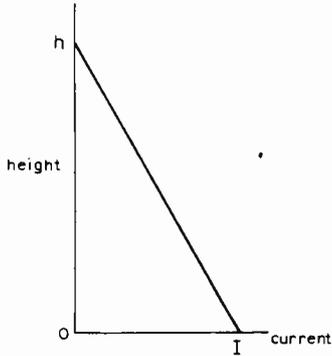


Fig.2. Current distribution in monopole.

$\cos \theta$, where θ is the angle of "take off" of the wave relative to the horizontal. The r.m.s. current distribution in such a monopole varies from a maximum, I , at the base, to zero at the top, as shown in Fig. 2. The radiated field strength is proportional to the average current multiplied by the height of the aerial, i.e. $Ih/2$. It is customary to regard this as $h_e \times I$ where $h_e = h/2$, and is called the effective height or radiation height of the aerial. The ratio of effective height to physical height h_e/h , the aerial form factor, is equal to $1/2$ in the case of an electrically short monopole, compared with $2/\pi$ or 0.636 for a quarter-wave aerial.

The radiation resistance, R_r , of an aerial is that resistance which, multiplied by the square of the aerial current, I , gives the power radiated: $R_r = 160\pi^2 \times h_e^2 / \lambda^2$ ohms. At a frequency of, say, 15kHz λ is 20,000m, and the radiation resistance of a 300m monopole would be 0.09Ω .

Many circuit designers are more familiar with artificial or dummy aeri- als than with the aeri- als themselves. Both receiver and transmitter specifications are written in terms of performance from, or into, a specified artificial aerial. Fig. 3(a) shows the circuit which

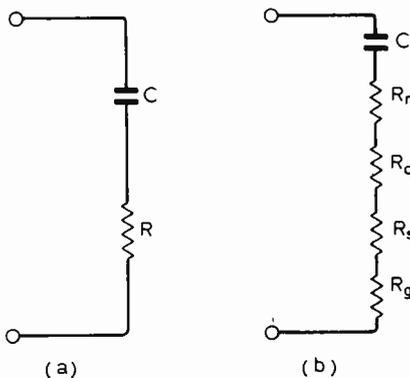


Fig.3. Simulation of v.l.f. aerial.

simulates a v.l.f. aerial. C will be around 1.02 to $0.25 \mu F$, and R about 0.1 to 1.0 ohms. In Fig. 3(b), R has been split into its component parts: R_r represents the radiation resistance; R_c is the effective aerial conductor resistance; R_s is an equivalent series resistance representing the losses in the aerial insulators, corona, masts and stays, and in the soil close to the aerial; and R_g represents the losses in the ground and earth connection. The power radiated is $I^2 R_r$, so the percentage efficiency of the aerial itself is $100\% \times R_r / (R_r + R_c + R_s + R_g)$.

The monopole considered earlier had a radiation resistance of only 0.09Ω , so you can see that the efficiency of this type of aerial is likely to be very low. Ground losses are usually minimized by using buried radial conductors, and conductor losses can be minimized by using numbers of conductors in parallel. The resistance of solid conductors is largely determined by skin effect, which, for copper, results in a skin depth of about 0.5mm at 15kHz . Conductivity is therefore proportional to the diameter of the conductor, so that doubling the conductivity of a given conductor results in a fourfold increase in weight and material cost, so that economic considerations play a large part in the lives of v.l.f. aerial designers.

Aerial efficiency can be considerably improved by raising the aerial form factor, such that the effective height more nearly approaches the physical height. This may be achieved by adding horizontal conductors to the top of the monopole, as in the inverted L and T type of aerial. In the case of the v.l.f. aerial it is more usual to find a large network of conductors connected to the top of the monopole. By this means it is possible to increase the aerial form factor from 0.5 to about 0.7 , which doubles the radiation resistance. A further advantage of the "top hat" of the aerial is that its capacitance is increased, resulting in a much lower voltage across the aerial insulators. Sometimes the top capacitance is in the form of a symmetrical network of conductors radiating from the top of the vertical system, as in the case of the NATO transmitter⁶ at Anthorn in Cumberland. This type of aerial is referred to as an umbrella aerial, due to its similarity to an umbrella frame.

Aerial efficiency has been defined as the ratio of radiation resistance to the total resistance of the aerial-earth system, but it must be remembered that the aerial needs to be tuned to present a substantially non-reactive load to the transmitter. One possible circuit configuration is shown in Fig. 4. The current which flows in the aerial tuning inductor (a.t.i.) is clearly equal to the aerial current and the power in the loss resistance of the a.t.i. can be considerable. However, by using a capacitance top to the aerial, its total capacitance is increased, so that the inductance required in the a.t.i. is decreased, resulting in a smaller loss

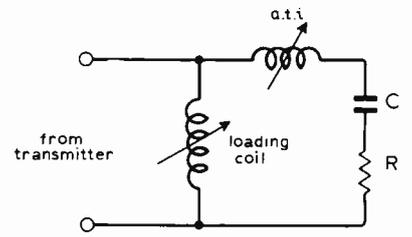


Fig.4. Aerial tuning circuit.

resistance and hence the total power loss is reduced.

An approximation to the static capacitance of the aerial may be made by regarding the aerial as a parallel-plate capacitor and making an allowance for the fringe effects. The area of the top plate of the capacitor is easily calculated, and the fringe effect may be allowed for by adding an area equal to the perimeter of the top hat multiplied by the average height of that perimeter⁷.

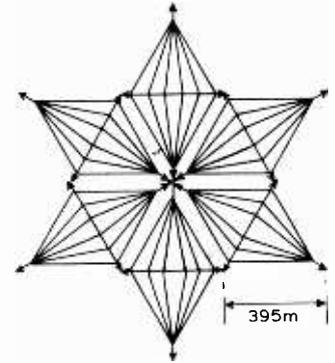


Fig.5. "Top hat" of Anthorn aerial.

An example will make this clearer. Fig. 5 shows the plan view of the top hat of the Anthorn aerial⁸, which has an average height of 183m . The total area of the star-shaped top hat is approximately $8 \times 10^5 \text{m}^2$, and the perimeter is $12 \times 395 = 4740\text{m}$. The effective extra area due to fringing is therefore $4740 \times 183 = 8.7 \times 10^5 \text{m}^2$. The total area upon which the capacitance must be calculated is therefore $(8 \times 10^5) + (8.7 \times 10^5) = 16.7 \times 10^5 \text{m}^2$. Capacitance of a parallel-plate capacitor is $A \epsilon_0 / d$, where A is the plate area in m^2 , d is the distance between plates, and ϵ_0 is the permittivity of free space (8.854×10^{-12}). The aerial capacitance is therefore 81nF . The capacitance quoted in the reference is 95nF but this includes the capacitance of the cage forming the vertical part of the aerial, and the capacitance of the aerial insulators.

It will be apparent from Fig. 4 that the current in the aerial must return to the transmitter via the earth connection and therefore the resistance of the earth system must be as low as possible in the interests of efficiency. High conductivity ground may be selected for the aerial site, as has been done in the case of the NATO v.l.f. transmitter at Anthorn, where the site is at the extremity of a lowlying peninsula on the Solway Firth.

To improve the ground conductivity still further, copper wires of 4mm diameter, buried to a depth of 30cm, radiate from the centre of the aerial system at 2° intervals, and extend over the whole area covered by the aerial. Alternate conductors, i.e. those at 4° intervals, extend to the edge of the site.

Mechanical construction

The usual method of supporting the aerial system is by means of guyed masts, which are cheaper than self-supporting towers. Lattice masts with a triangular cross-section are normally used, as this shape provides anchoring points 120° apart for the stays, which are usually fitted at heights of every 50m or so.

The conductors which form the aerial down-lead and top capacity are usually steel-cored aluminium cables of about 2.5 to 4cm diameter, similar to those used for overhead electric power transmission. This diameter cable is necessary to prevent corona, rather than for its current carrying ability. Anti-corona rings and insulators are fitted at the support points. It is interesting to note that at v.l.f. the onset of corona occurs at a voltage about 17% lower than on 50Hz power systems.

Some unusual aerials

The aerials so far mentioned have been fairly conventional in design, but there have been some unusual systems.

One of these was the German Goliath transmitting station built in 1943 near Kalbe on the marshes of the river Milde (ref. 7, p.144). This aerial used three identical interconnected hexagonal shaped top capacitances, shown in Fig. 6. Each hexagon had an insulated tubular centre mast which was tuned with its own a.t.i., and a fourth a.t.i. was connected between the transmitter output and the common connection to the six up leads at the centre of the system.

By using this multiple tuning technique a very high overall efficiency was achieved, resulting in a radiated power varying from 300kW at 15kHz to 900kW at 60kHz.

The chief reason for the improved efficiency of the multiple tuned aerial is that the earth current is shared by the earth systems. Suppose the system consists of n aerials, each being tuned and carrying a current of I amps. If the earth resistance were R_g ohms, the total power loss would be nI^2R_g watts. If only one aerial were used, having the same earth resistance R_g , but with a current of nI amps, then the power loss would become $(nI)^2R_g$ watts, which is n times as great as with multiple tuning. The earth losses are therefore reduced in direct proportion to the number of multiple sections.

A similar technique has been used in the U.S. Navy station at Cutler, Maine, where the aerial consists of two separate star-shaped umbrella systems, each

	Rugby	Anthorn	Goliath	Cutler	Tsushima
Number of masts	12	13	18	26	1
Average height (m)	250	200	175	258	450
Site area (km ²)	3.0	2.5	2.6	9.5	0.13
Static capacitance (°F)	80	95	115	225	34.4
Effective height (m)	133	112	137	145	200
Radiation resistance (mΩ)	80	56	85	100	73
Q of system	—	—	500	365	470
Normal working voltage (kV r.m.s.)	110	120	180	210	170
Power output (kW)	450	550	700	2000	135
Radiated power (kW)	65	50-100	350	1550	10
	at	from	at	at	at
	16kHz	16 to	16kHz	14.7kHz	10.2kHz
		20kHz			
Percentage efficiency	14.4	9.1	50	77.5	7.6
	at	at	at	at	at
	16kHz	16kHz	16kHz	14.7kHz	10.2kHz

similar to the Anthorn aerial but larger. Each system has its own tuning inductors and they are normally operated in parallel.

Hills and valleys have also been used for v.l.f. aerials. The Post Office Station at Criggion, built during the Second World War as a standby for Rugby, originally used only three masts, the other two supports for the aerial array being situated on adjacent hill tops. Valley span aerials are also used, in which the transmitter is located in the dip between the hills and the top capacitance spans the valley. Such a system has been planned for the Omega station in Hawaii.

Japan's Omega station

The Omega transmitter being built by Japan¹⁰ is situated by a bay in the north-west corner of the island of Tsushima in the Korea Strait. Its aerial is unique in that it employs a single insulated tubular mast, 450 metres high, with a top capacitance consisting of 16 4cm diameter conductors radiating outwards and downwards from the top of the mast. The active length of each top conductor is about 315 metres. The earth system consists of 90 radial conductors, each about 200 metres long, spaced at 4° intervals.

The mast itself is about three metres diameter, and is made up of 68 cylindrical sections. The weights of the sections vary from about 19 tons for the lower units to 16 tons for the upper ones. Stays are spaced approximately 120° apart, and fitted at six levels. Each stay is broken up into separate lengths by means of six insulators.

The whole assembly is supported on an insulator unit, comprising six insu-

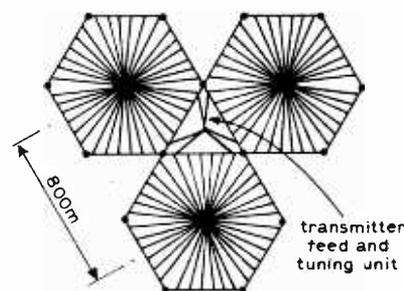


Fig.6. "Top hat" of Goliath aerial.

lators, which is designed to stand a maximum working load of 3300 tons.

Typical aerials

The major physical and electrical characteristics of some of the aerials mentioned are listed in the table. For comparison, the effective height, radiation resistance, and Q of the communication aerials have been normalized to a frequency of 16kHz. Performance figures for the Omega station are based on the scale model described in reference 10 and they refer to a frequency of 10.2kHz.

Figures marked "Q of system" are approximate and refer to the Q of the aerial earth system plus the a.t.i. The effective Q of the aerial circuit under working conditions will be less than the figure quoted due to the output impedance of the transmitter.

Efficiency is tabulated as the ratio of radiated power to transmitter output power. This is not necessarily the maximum possible efficiency, as damping resistors are sometimes included to widen the bandwidth.

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Research Notes

E.E.G. test for telepathy?

Research on extra-sensory perception at Stanford Research Institute, California, has produced what looks like a useful screening process for picking out people who possess telepathic powers without themselves being aware of the fact. It results from a chance observation made during an experiment which failed. In this experiment one subject, the "home" subject, had a light flashed in his eyes at frequencies of 6 or 16 flashes per second. This tends to lock the alpha rhythm (normally about 10Hz). If by telepathy the "remote" subject's alpha rhythm were also locked the e.e.g. would reveal the effect. In fact no such effect was observed, but in a few remote subjects the alpha rhythm was depressed in amplitude while the light was flashing. The remote subjects who showed this effect had no consciousness of the state of the light. When asked to say when it was flashing they did no better than chance.

Better propagation forecasts?

A new way of predicting solar activity has been discovered by G. M. Brown of University College of Wales, Aberystwyth. It stems from his observation that there is a strong correlation between the sun's effect on the earth's magnetic field and the number of sunspots six years later. The reason for this is not known, but it holds good over a time span which goes back to 1885, and the correlation is very close indeed. If it proves to be a genuine effect and not a freak of statistics then it could give radio propagation experts a valuable method of improving their short-wave propagation predictions.

The magnetic effect in question operates on the horizontal component of the earth's field. This normally goes through a minimum at about 11.00 hours local time, but on "Abnormal Quiet Days" (AQDs) the minimum is at some other time. It is the AQDs which predict the sunspot numbers. Since AQDs are most frequent at sunspot

minima it could be that they mark the beginning of new cycles of solar activity rather than the end of old ones. "If this relationship proves valid it implies that the sun 'breathes' with an 11-year-period, such that the size of a solar activity maximum is determined at the very beginning of a cycle, or, perhaps the very end of the preceding cycle, from the "depth" of the solar minimum."

Wave power looks good

The subject of a piece of writing which contains the words wavelength, bandwidth, matching, resonant, tuned, and reciprocity might be expected to be some sort of electrical circuit. Yet all these terms (and others familiar to the electrical or electronic engineer) are to be found in a letter to *Nature* (vol. 254, p504) about a purely mechanical device. It is a gadget for extracting the energy from ocean waves. First proposed by S. Salter of the Mechanical Engineering Department at Edinburgh University, the "Salter cam" has now been investigated by the Marchwood Engineering Laboratories of the CEGB. The Salter cam is a floating boom whose cross section is roughly the shape of a duck's body. As a wave passes beneath it, the cam is given a twist. This exploitation of the twisting part of the wave's energy is the important element in the design. The question is: how much wave energy is extracted? The Marchwood results show that with exact matching of the dimensions of the cam to the wavelength the efficiency can exceed 90%, while the 3dB bandwidth covers about one octave. These results have been obtained using small cams in a test tank, and have yet to be confirmed with full sized devices in real sea conditions, but certain losses in the system are frequency-dependent and should be less for full-sized ocean waves.

Some means of transferring the energy to the shore will have to be developed. Possibly the motion of the cams could be used to pump water.

A useful consequence of the efficiency of the cam is that the water behind it is calm, most of the wave energy having been extracted. So the device seems to have possibilities as a floating breakwater.

Digital filters reveal weather trends

Statisticians have long used a mathematical device, the moving average, for removing the short-term peaks and troughs from a time-series of data and so revealing the underlying trend. The "moving average" process is analogous to a low-pass filter, which removes high-frequency noise and lets pass a low-frequency signal.

Information stored in electrical form can be passed through a real low-pass

filter. Scandinavian research workers have been using this trick to very good effect to smooth away the short-term fluctuations from climatological data. The data go back about 1 000 years and were smoothed by means of a digital low-pass filter whose cut-off was set at various intervals up to 200 years. The results reveal an unexplained correlation between the climate in England and that in Greenland, with a time-lag of 286 years. On this basis, the further outlook in Britain is colder.

Ear temperature clue to brain-damage

Measurements taken with the aid of a thermistor probe suggest that the temperature just inside a newborn baby's ear may provide a diagnostic test for brain damage. When doctors at the London Hospital Medical School compared the temperature of the ear of a newborn baby with that of the oesophagus they found that the ear temperature was consistently 0.2deg.C higher. This is not surprising in view of the discovery that in a normal baby 70% of the body heat is generated in the brain, compared with less than 20% in a resting adult. It is hoped that ear temperatures may vary from the norm in brain-damaged babies and so provide a simple means of diagnosis.

Holes in the ionosphere to aid radio astronomy?

The exhausts from large space-rockets make holes in the ionosphere. The release of large quantities of gas mops up free electrons and so reduces ionization. The Spacelab launch made an ionospheric hole which persisted for several hours.

A suggestion has been made that such a hole, in the right place, would permit radio astronomy at frequencies which normally do not reach the earth.

Towards the 12GHz consumer f.e.t.

When direct TV broadcasting from satellites begins there will be a need for cheap active devices for use in home receivers. In Japan, a process for making gallium arsenide microwave f.e.t.s is under development. Only one mask is required. This creates titanium gate contact strips 3 microns wide, on a surface layer of p-type gallium aluminium arsenide grown on the n-type layer of GaAs which forms the channel. A etching process removes unwanted gate material and undercuts the base contact strips by 1 micron. The undercut strips act as umbrellas to shield the channel material from the subsequent metalisation processes which create the source and drain contacts.

A digital waveform synthesizer

Variable a.f. function generator

by R. A. J. Youngson

The output of the waveform generator is a series of consecutive, negative-going, voltages of equal duration and of independently variable level. Thus complex waveforms can be synthesized.

The circuit described (see Fig 2) divides each cycle into 32 bits providing control of up to the fifteenth harmonic. Fig. 1 shows the synthesis of a sinusoid. The amplitude A of each pulse is adjusted to form an approximation to the curve required.

Circuit

Essentially, the circuit is designed to decode a series of five-digit binary numbers to a one-of-thirty-two output. A square-wave oscillator, using a Schmitt trigger (1/2 SN7413) and a few external components, drives a chain of six binary dividers (one SN7493 and one SN7473). The purpose of the first divider is to ensure that the mark-space ratio of the input square-wave is 1:1, otherwise a variation in the output pulse width would result. The Q outputs of the next four divider stages

are connected to the A, B, C, D inputs of two SN74154 four-bit-binary to one-of-sixteen decoders. These inputs are common to both decoders. The Q and Q-bar outputs of the last divider stage are coupled to the strobe inputs of the two decoders as shown in Fig. 2. This arrangement ensures that the decoders operate alternately as the states of the outputs of the last divider change. The normal state of the decoder outputs is logic 1 and for any output this is switched to 0 when the binary code corresponding to that output is applied to the A, B, C, D inputs. There are two strobe inputs in each decoder. If either or both of these is at logic 1, then all the outputs will be at 1 regardless of the state of the A, B, C, D inputs. Only if both of the strobe inputs are grounded will the device perform the decoding function. Fig. 3 shows a truth table for the switching of the decoders.

Each of the thirty-two outputs is grounded through a 20kΩ potentiometer, the moving contact of each being connected to an output bus via a

100kΩ isolating resistor. In the logic 1 state, the outputs of the decoder are at about +4V but because of the large number of paralleled resistors, this is dropped to between 150 and 200mV, still allowing an adequate signal to noise ratio. The output impedance lies between 3 and 4kΩ.

Since there are six stages in the divider chain, the output frequency is

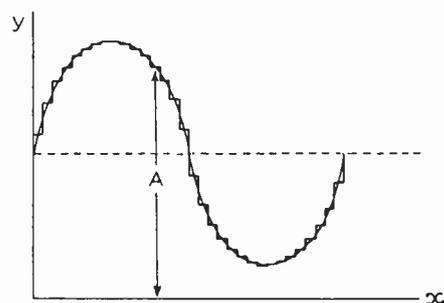
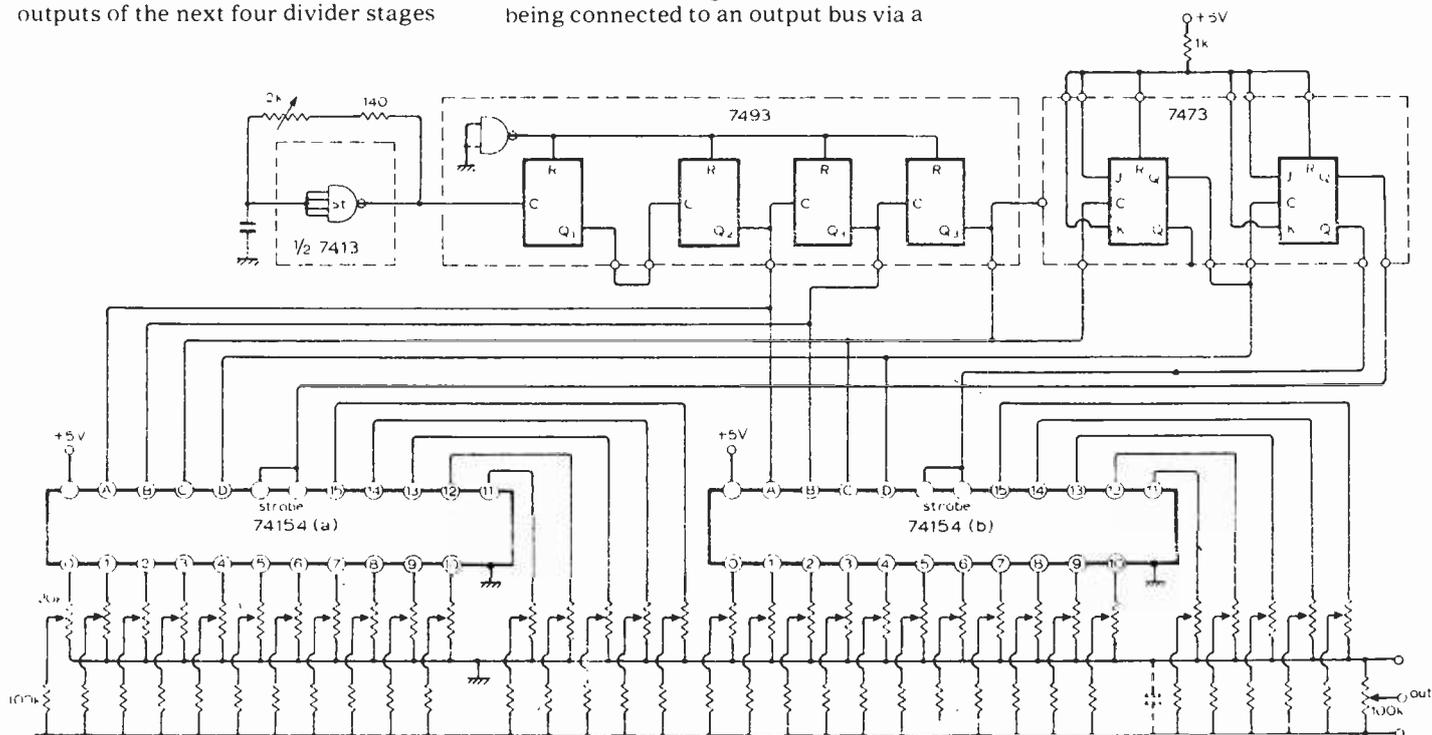


Fig.1. Synthesis of a sinusoidal waveform.

Fig.2. Circuit diagram of synthesizer.



only 1/64 of the input frequency, so the oscillator has to run at up to about 1MHz to achieve an output frequency range of from a fraction of a hertz to about 20kHz. The clock oscillator described is capable of covering about four octaves with a single capacitor. If a second capacitor, of exactly equal value, is switched in parallel, the frequency drops by an octave—a useful characteristic in a musical instrument.

The t.t.l. short rise and fall times permit a precise output waveform with no easily detectable gaps between pulses. In addition, the frequency-determining elements should be stabilized against temperature or voltage changes. The levels of all the outputs from the decoders are uniform, permitting accurate calibration of the 32 controls.

The circuit requires only one 5V supply, but this must be well stabilized. This is most conveniently accomplished by using an i.c. regulator. The current consumption of the whole circuit is less than 140mA.

The output waveform is most readily appreciated on an oscilloscope. For some purposes a stepped waveform is inconvenient, and a 0.01 μ F capacitor wired across the output will filter out the steps to give a smooth waveform. This value is sufficient at most audio frequencies but a larger value is required (i.e., 1 or 2 μ F) if the device is used as an envelope generator at very low frequencies.

Construction

The main circuitry is conveniently fitted on 0.1in pitch Veroboard. It is best to arrange the two decoders so that all the outputs can be taken from one edge of the panel in the correct order. The miniature potentiometers used were mounted in two rows of 16, an inch apart, but an alternative arrangement may be preferred. Colour-coded wires are recommended for the connections between the circuit board and the control panel.

Applications

The original purpose of the device was as a generator of musical waveforms but there are wider applications. When used as a musical tone synthesizer, waveform patterns, such as photographs of oscilloscope tracings, may be copied. Due observance of fine detail will be found to be important. One should, however, bear in mind that the waveform is not the only parameter that governs the sound.

The device may be used as an envelope generator to control a voltage-controlled amplifier supplied with tones from a further waveform generator, so producing accurate attack and decay curves each time a note is played. It can be arranged that the first half of the generator cycle is used for the attack curve and the second for the decay curve.

DIVIDER OUTPUTS		DECODER (a)	DECODER (b)
Q	\bar{Q}		
1	0	all outputs 1	decoding
0	1	decoding	all outputs 1

Fig.3. Truth table for decoder switching.

The form of the output is, of course, entirely independent of frequency and a third application is the use as a rhythm generator by appropriately reducing the clock frequency. A sequence of up to 32 beats in a bar may be produced, the loudness as well as the presence or absence of each being independently controllable. Realistic effects are possible ranging from single beats to a drumroll.

Yet another application is as a voltage source for a voltage-controlled oscillator so as to produce a sequence of 32 controllable pitches. The device could, for instance, be programmed to play the bass notes of a short musical composition.

Acknowledgement: The author is indebted to earlier workers who evolved the basic ideas on which this device operates.^{1,2}

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What are Circards?

Circards are a unique way of collating and presenting data about circuits in a compact and easily retrievable form. The sets of 203 x 127mm (8 x 5in) double-sided cards are designed for easy filing in standard boxes and for easy access at the desk or at the bench, where transparent plastic wallets keep the cards in good condition.

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How speech can be compressed and expanded

Methods of speeding or slowing speech reproduction without distorting nuances or intelligibility

by S. L. Silver

In past years, variable speech-control devices employing electromechanical techniques have been commercially available as an auxiliary unit for reel-to-reel tape machines. Recently, however, with the advances in sophisticated electronics and complex solid-state circuitry, rate-variable equipment has been produced as a built-in facility in cassette machines for the time compression and expansion of recorded speech.

In our speech-oriented technology, the recorded voice is becoming increasingly significant in the transfer of information. But owing to the physiological limitations of the organs of speech, we lack the ability to produce speech as rapidly as the ear-brain perceptual mechanism can process it. Normally, average speaking rates range from about 110 to 175 words per minute, whereas the capacity of the human auditory system to absorb speech in real time is more than twice that rate. It follows, therefore, that in order to cope with the proliferation of audio data, we should take advantage of our higher listening-rate capability by utilizing some form of rate-controlled speech processing.

The simplest approach is to reproduce tape-recorded speech at a faster rate than it was originally recorded. This may be feasible for small speed-up ratios, but larger variations (in the order of 50 percent) would result in serious deterioration of quality and intelligibility. What happens, of course, is that the speech spectrum is proportionately scaled-up in time, thus distorting the normal voice pitch. One solution to the problem lies in compressing the recorded speech signals as a function of time during playback, so as to accelerate the speaking rate without altering the pitch or tonal quality of the voice. Conversely, the speech rate could be slowed down by expanding the recorded signals in a pitch-invariant manner.

With these methods, the listener is allowed to control the playback rate of audio presentation and, hence, select his optimum listening speed without sacrificing intelligibility and comprehension.

Potential applications

Although rate-controlled speech has already proved its usefulness in some fields, modern methods have opened up

a wide range of potential applications. For example, listening to time-compressed speech can provide an efficient learning medium for individuals with special educational problems, such as the visually handicapped. With high-speed listening, recorded conference proceedings can be reviewed rapidly, since more subject matter is reproduced in a given time.

In transcribing recorded speech directly to the typewriter, the word rate can be adjusted according to the individual's typing ability. Also, programme material for a radio broadcast transmission can be played back within the desired time limit, without the need for editing the recorded text.

During expansion, recorded speech can be slowed down to a convenient word rate for teaching retarded children, and in speech therapy in general. Low-speed listening can also be helpful in learning a foreign language, where the listener is able to set his own listening pace, making it easier to imitate the articulatory gestures of speech production. Similarly, in linguistic studies, a low word rate enables the listener to make more precise phonetic transcriptions.

In order to gain better insight into the technical aspects of compressed and

expanded speech in time, one must appreciate the dynamic and phonetic characteristics of vocal sounds. Therefore, before considering the methodology of rate-altered speech, it would be useful to explore briefly some of the fundamental features of the speech process in acoustical terms.

Speech production

Speech signals consist of rapid fluctuations in air pressure, wherein sound energy is generated and radiated by the vocal system. A schematic representation of the voice-producing mechanism is shown in Fig. 1. A column of air is expelled from the lungs, passes through the glottal aperture, and drives the vocal cords into forced oscillation. The vocal cords, acting like an acousto-mechanical relaxation oscillator, chop up, or modulate the air stream into near-periodic sawtooth pulses. These time-pressure patterns, in turn, excite the vocal tract where the position of the articulators (tongue, lips, palate, etc.) establish certain resonant conditions.

Sample-and-discard methods

Human speech may be regarded as the process of transforming a message from sequence of basic sound units — the phonemes — into a continuous acoustical signal that conveys information to the listener. Phonemes are considered

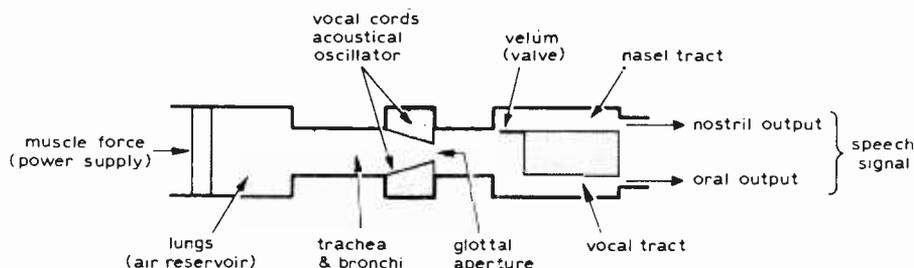


Fig. 1. Schematic representation of the human vocal system.

to be the minimum recognizable speech components that can be distinguished from any other speech sounds produced. Natural connected speech, however, does not merely consist of isolated sound elements or discrete events. The acoustical attributes of each sound unit constantly change the flowing speech pattern due to the merging influence, or interaction, between adjacent phonemes. In this process, phonemes are produced at the rate of about 10 to 20 per second, the duration of a single sound element ranging up to 100ms.

Fortunately, the time interval of the average phoneme in continuous speech exceeds the minimum duration necessary for intelligible perception by the listener. The temporal redundancy of vocal communication, in effect, is sufficiently high to permit direct manipulation of the time-scale of a complex waveform. This is an important factor when considering the sample-and-discard method of speech processing.

In an experimental demonstration of speech compression, Garvey¹² edited recorded voice tapes in such a way as to produce speech sounds with a high word rate. Using a manual sampling technique, he carefully cut the tape into periodic sections (independently of the speech content), discarded tiny interspersed segments, and spliced the remnants together. He found that the shortened composite tape, when reproduced at normal speed, retained most of the original voice quality.

Subjective tests were then carried out to determine the effects of increased compression rates on intelligibility. The results are illustrated by the curve plotted in Fig. 2. Here the compression rate, expressed in percentage, is defined as the ratio of the playback speed to the normal recording speed, e.g., a compression rate of 100 percent corresponding to normal reproducing speed. Speech compression at 200 percent indicates that reproduction is twice as fast, so that playback time is 50 percent of recording time. Clearly, the curve shows that intelligibility is remarkably good at compression rates up to 200 percent after which it declines sharply to 400 percent, where it becomes extremely poor.

It should be pointed out that a distinction must be drawn between the intelligibility of single words and the comprehension of complex information. A word is considered intelligible if the listener can recognize it and repeat it, but comprehension signifies the retention of ideas in which specific facts can be subsequently recalled. Generally, as the word rate of compressed speech is accelerated, comprehension declines more rapidly than intelligibility. Another point of consideration is that the degree to which speech sounds can be speeded up without loss of listening comprehension is highly dependent upon the difficulty of the recorded passages and the familiarity of the listener with the subject matter.

Electromechanical techniques

Obviously, the manual sample-and-discard method is time-consuming and impractical, and it would be necessary to implement the control of speech rate by automatic means. It is possible to perform time compression by the elimination of those pauses between words that exceed a preset time interval. Pause suppression can be accomplished by means of a fast-acting stop-start clutch, which automatically stops the tape player during the pause interval and restarts it in the presence of speech signals.

Unfortunately, such a system ignores the fact that speech is not merely a chain of words randomly linked together, but is structured by grammatical rules to transmit complex data to the hearing mechanism. In this process, the pauses (as well as the stress levels, intonation patterns, and inflections) introduced by the talker are used by the listener to decode these complex ideas. Accordingly, the natural pause-to-speech sound relationship at the juncture of phrases and sentences, and even the intermittent hesitation pauses between words, should be maintained for maximum listening comprehension.

In order to preserve the relative rhythm of the original recorded speech sounds during speech processing, time-controlled speech may be achieved by automatically sampling the signals in

periodic segments. About thirty years ago, Gabor³ proposed an electromechanical scheme for compressing speech by implanting a magnetic head assembly in a rotating wheel and scanning a magnetic tape.

In the 1950s, Fairbanks *et al.*⁴ developed apparatus employing multiple pickup heads set in a revolving drum operating in conjunction with a magnetic tape loop. The differential tape-to-drum speeds could be varied independently to attain any degree of compression or expansion. Subsequently, the Acoustical Time Regulator designed by Springer,⁵ utilized a rotary head assembly capable of direct reproduction because of synchronization between tap speed and head rotational speed. Further development of this device produced the Information Rate Changer, which combined the functions of pitch and tempo regulation.

Rotary-head processor

Referring to Fig. 3 the rotary-head processor incorporates four separate playback heads mounted in quadrature, with the tape wrapped around the head assembly for one-quarter of the perimeter. Signal outputs from all heads are connected in series and passed through a slip-ring arrangement, then fed to a pre-amplifier in a conventional way. In the tempo-change function, the rotational speed of the playback heads is linked in such a way with the rotational speed of the capstan, i.e., the linear speed of the tape, that a constant pitch is maintained over a wide range of speeds.

In the compression mode, the playback time is shortened by skipping individual segments of the taped speech. Here the head assembly rotates in the same direction as tape travel, with the tape-to-head gap velocity equal to the speed at which the tape was recorded. The absolute tape velocity (with respect to the tape deck) determines the time duration of the compressed speech. During this operation, one head gap leaves the tape-contact area while an arriving head establishes contact. However, at the instant of transition, a definite interval will recur when the tape-segment signal between the head gaps will be effectively omitted, thus contributing nothing to the reproduced output.

In the expansion mode, the playback time is lengthened by repeating individual segments of the recorded tape. Here the head assembly rotates in a direction opposite to tape action, while maintaining a constant relative tape velocity equal to the recorded speed. In this case, two head gaps establish tape contact almost simultaneously; one head gap picks up an individual tape segment as the next head gap reproduces the same segment. In order to prevent the occurrence of detectable distortion, the repetitive interval during expansion (or the discard interval

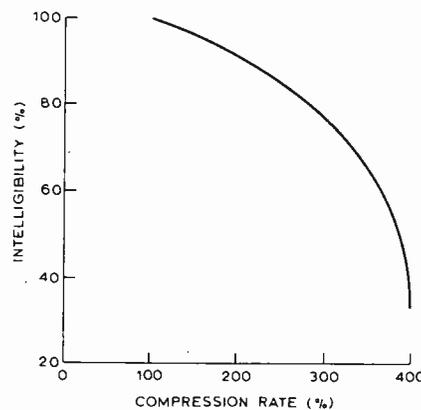


Fig. 2. Speech intelligibility versus compression rate curve (after Garvey).

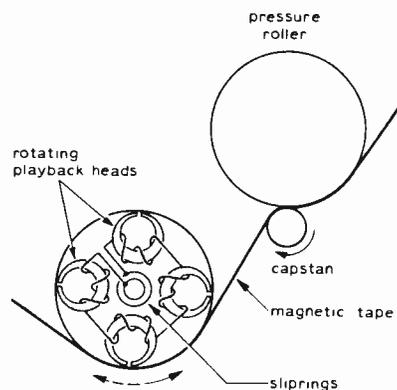


Fig. 3. Schematic view of the rotary head assembly (after Springer).

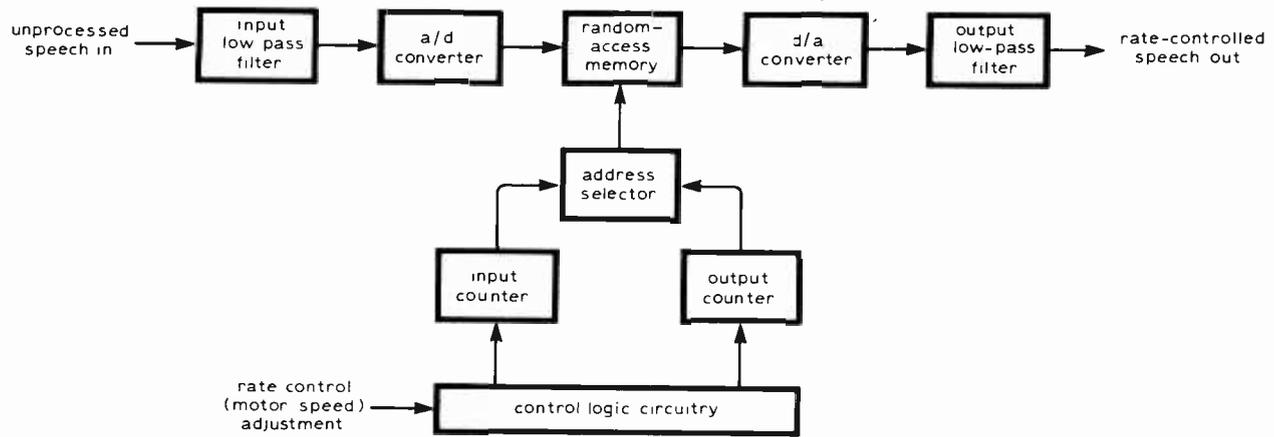


Fig. 4. Simplified block diagram of a speech processor using electronic sampling techniques.

during compression) must be shorter than the average basic sound unit. For this reason, the segmented lengths are fixed precisely in time, say 30ms, which is considered below the perception threshold of audible disturbances.

In the pitch-change function, it is possible to raise or lower the pitch simultaneously with, or independently of, the tempo. Under these conditions, rotating the head assembly in the direction of tape motion and decreasing the relative tape-to-head gap velocity will result in a lowering of the pitch. Rotating the heads in a reverse direction will have the opposite effect. The pitch-change function may have potential value as an aid for individuals with certain types of hearing impairment. For example, by shifting the voice spectrum down by some factor, say one octave, in real time, speech sounds can be transposed within the range of hearing of partially deaf persons.

Electronic techniques

In contrast to electromechanical systems, speech compression can be performed electronically without discarding discrete portions of the input signal. An example of this approach is the Harmonic Compressor⁶ which uses spectral analysis and synthesis to produce a pitch-normalized output. Initially, the speech signal is applied to a bank of 36 bandpass filters (covering the range from 200 to 7,400Hz) which separate the signal components into 200-Hz bands. The filtered output of each channel, in turn is fed to its corresponding frequency divider, thus halving the frequencies of the narrow-band signals. The output from all channels is finally combined in a summing network, and filtered to remove the distortion components.

To restore the half-frequency signals to their normal values, the synthesized speech is recorded on tape and then reproduced at double speed. The word rate is now twice the normal rate, with the voice pitch normalized with the original speech signal. Although the Harmonic Compressor is restricted to a compression factor of two, the principle, of course, is applicable to other

speed-up ratios. Speech expansion may be accomplished by frequency multiplication instead of division.

At present, time compression/expansion of speech is being implemented with digital processing using electronic sampling techniques.⁷ The sampling theorem states that if the sampling rate is a minimum of twice the highest frequency components of a continuous signal, the sampled version of that signal can be converted back to its original form. Accordingly, the speech signals to be processed are low-pass filtered at the input to ensure that there is no acoustic energy above the maximum frequency of interest, then applied to an analogue-to-digital converter at the appropriate sampling rate. If, for example, the desired upper frequency limit of the filtered signal is 7kHz, and the speech to be compressed is speeded-up by a factor of 2.5, then the sampling rate is required to be a maximum of 35kHz.

In the electronic speech processor shown in Fig. 4, the input counter stores the sampled signals in successive locations of the random-access memory (r.a.m.), with the final location followed by the initial one. The stored samples are retrieved by the output counter (at a fixed rate) from consecutive locations of the memory. Finally, the data is converted back to analogue form, then low-pass filtered at the output to reconstruct a rate-altered version of the original speech signal.

In effect, the r.a.m. provides a means for sequential storage and presentation of signal samples, and, since the frequencies of the original signal have been restored with respect to memory space, they are reproduced at their normalized values. The relationship between the variable read-in rate and the constant read-out rate determines the compression and expansion ratios of the pitch-corrected output. Thus, if the read-out rate from the memory is made variable, the speech processor can function as a pitch-changer.

In a commercially available cassette

machine which incorporates a digital signal processor, the basic assembly is made up of printed circuit cards. In the playback mode, the listening speed can be varied continuously (with a single control) from 0.5 to 2.5 the normal recording speed, without altering the pitch characteristics of the signal. Tape speed automatically reverts to normal when the machine operates in the record mode. The transport mechanism employs servo-regulated speed control with tachometer feed-back to establish a linear relationship between read-in rate and capstan motor speed. Also, an input facility is provided to compress or expand audio signals from an external variable-speed source.

Another electronic sampling approach to speech processing makes use of analogue shift registers, commonly known as bucket-brigade devices. These units, however, are still undergoing development, the main objective being to integrate the basic system on a low-power chip.

This article has been reprinted from the American sound engineering magazine *db*.

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Sorting out signs

Sign conventions in electric and magnetic circuits

by A. T. Morgan, M.Sc., M.I.E.E., M.I.Mech.E.

Middlesex Polytechnic

To correctly interpret the results of calculations about electrical networks, it is important to be logical and consistent. This article outlines a logical system using conventional methods of treating circuit and phasor diagrams.

The matter of notations and sign conventions is one which often causes confusion. For example, one often reads the statement that induced e.m.f. in an inductance coil is $-Ldi/dt$ without any indication of which is the positive direction of applied voltage, induced e.m.f. or current! Also, from many transformer circuit diagrams, it is impossible to decide what the relative polarities of the two winding voltages are.

To write the equations that describe the behaviour of an electrical network and correctly interpret the results of calculations, it is essential to make clear a logical and consistent system of notations and sign conventions.

Scroggie* has pointed out some of the difficulties and confusions that may be encountered and has suggested a new and interesting method of labelling currents and voltages in circuit diagrams and phasor diagrams. This article outlines a logical system using conventional methods of dealing with circuit and phasor diagrams.

Resistance circuit

Consider first the simplest possible circuit consisting of a single resistance with an applied voltage v . First choose a convention for the positive direction of v as shown by the arrow (arrow head positive) and for e_R , the voltage drop across the resistance, which is clearly the same voltage see Fig. 1(a). Now choose a positive direction for the current i as shown by the arrow. With the selected positive directions of

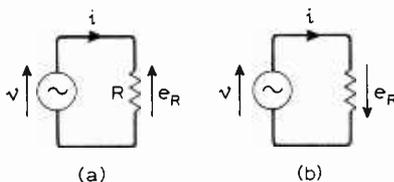


Fig. 1

voltage and current, a positive applied voltage would give a positive current. The power, which is the product of voltage and current, is positive i.e. passing from the source to the resistance. Also, as a positive applied voltage v would give a positive current i , then if v is sinusoidal then i will be sinusoidal and in phase with v , i.e. the power factor is unity and the phase angle zero, see Fig. 2(a). One can write

$$v = e_R = iR$$

where R is a positive quantity.

We use lower-case letters for instantaneous values of quantities and upper-case letters for phasors.

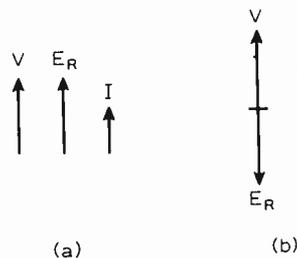


Fig. 2

Suppose that the positive voltage and current directions had been selected as shown in Fig. 1(b). Notice that in this case all voltage and current arrows point the same way round the loop. The applied voltage v and the resistor voltage e_R are now equal and opposite, i.e.

$$v + e_R = 0 \text{ or } v = -e_R.$$

$$\text{Also } v = iR \text{ and } e_R = -iR.$$

The phasor diagram is shown in Fig. 2(b).

You may find this convention for resistance strange. However it is quite common when considering an inductance to consider applied voltage and inductor voltage or "induced e.m.f." as

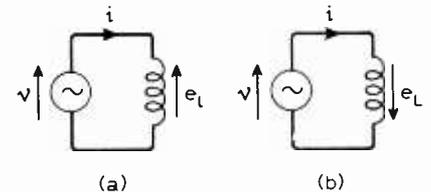


Fig. 3

equal and opposite. The above convention would thus make for consistency!

Notice that in this case the supply power $P(vi)$ is positive while the resistor power $P_R(e_Ri)$ is negative. Thus positive power is power generated and negative power is power dissipated. It is not necessary to know if a source or sink of power exists between any pair of terminals, as the algebraic sign of the power indicates whether power is being generated or dissipated. One can write the energy equation $P + P_R = 0$.

Inductance circuit

Consider the simple inductance circuit shown in Fig. 3. The sign conventions for voltage and current have been selected and Figs. 3(a) and (b) show two alternative choices for the positive direction of inductor voltage e_L . (Compare with Figs. 1(a) and (b).)

Consider first diagram (a). If the current is positive and increasing then the inductor voltage will be in the opposite direction to the current i.e. positive. Thus

$$e = +Ldi/dt.$$

Kirchhoff's law gives

$$v - e_L = 0, \text{ thus}$$

$$v = e_L = Ldi/dt.$$

Putting the operator d/dt equal to j for a.c. sinusoidal operation:

$$V = E_L = j\omega LI = X_L I, \text{ where } X_L = j\omega L.$$

The phasor diagram is as shown in Fig. 4(a). Notice that the expression for E_L

* Phasor Diagrams by M. G. Scroggie (Iliffe).

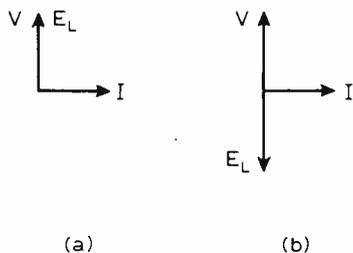


Fig. 4

above does not contain the usual minus sign. It is however correct.

Next, consider Fig. 3(b). If the current is positive and increasing the inductor voltage e_L which is in the opposite direction to the current, is now negative i.e. in the opposite direction to the new positive direction. Thus

$$e_L = -Ldi/dt \text{ or } E_L = -j \omega LI.$$

$$\text{Also } v + e_L = 0; \text{ so } v = -e_L.$$

The phasor diagram is shown in Fig. 4(b).

Notice that the expression for e_L now contains the customary minus sign. The minus sign is usually quoted but without any meaning as explained previously. Also notice that the statement sometimes made that the minus sign signifies that the voltage e_L opposes the current is incorrect. It is sometimes in the same direction as the current. It does of course oppose the change in current.

In the rest of this article the conventions for resistance and inductance are as shown in Figs. 1(a) and 3(a) i.e. with positive voltage and current arrows in opposite directions. This is not any better than other alternatives. All that is required is a statement of the conventions adopted followed by consistency in the use of them. Remember that at any pair of terminals, positive voltage and current arrows in the same direction means that positive power is power generated, whereas positive voltage and current arrows in opposite directions means that positive power is power dissipated.

RL circuit

Consideration of an RL series circuit is now straightforward. In Fig. 5(a),

$$v - e_R - e_L = 0$$

$$v = e_L + e_R$$

and the phasor diagram is below (b).

One often sees the phasor diagram for this circuit drawn as shown in Fig. 6(b).

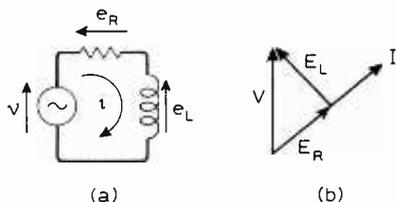


Fig. 5

This corresponds to the conventions shown in Fig. 6(a), although this is rarely stated. Notice that the positive direction of voltage e_L is taken to be the same way as positive current whereas positive resistor voltage is taken the opposite way to current.

$$v - e_R + e_L = 0$$

$$v = e_R + (-e_L)$$

where $e_R = +iR$ and $e_L = -Ldi/dt$. i.e. applied voltage is equal to resistor voltage plus a term equal and opposite to inductor voltage. This seems to be unnecessarily complicated.

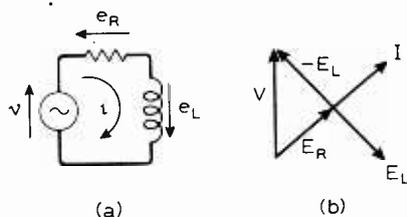


Fig. 6

Circuit with two way power flow

In Fig. 7 both boxes can act as either sources of power or sinks. In this case the power flow may be from left to right or vice versa. With the positive voltage directions as shown it would be equally logical to adopt either of the alternatives shown in Fig. 7 for positive current direction i.e. indicating positive power flow from left to right or vice versa.

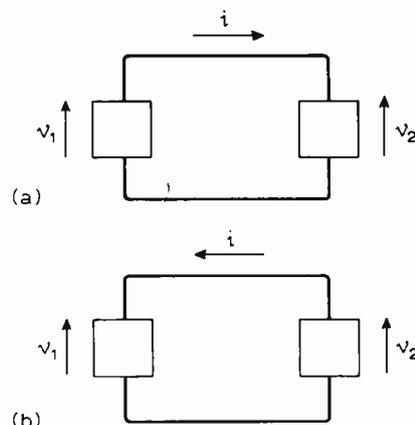


Fig. 7

Adopting alternative (a), positive power flow from left to right, and drawing the phasor diagram as shown in Fig. 8(a), the phase angle between current and voltage indicates the direction of power flow. If I has a component in phase with V_1 then the power flow is positive, i.e. from left to right. If I has a component in antiphase with V_1 as in Fig. 8(b) then the power flow is negative i.e. from right to left.

For an impedance Z between the two boxes, as shown in Fig. 9, the corresponding phasor diagrams will be as shown in Fig. 10. Equation representing Fig. 10(a) and (b) is $V_1 = V_2 + IZ$.

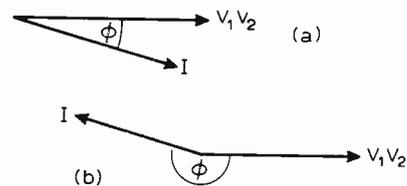


Fig. 8

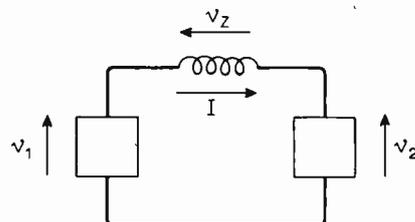
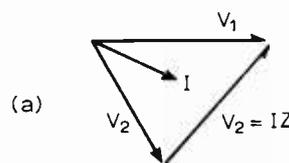
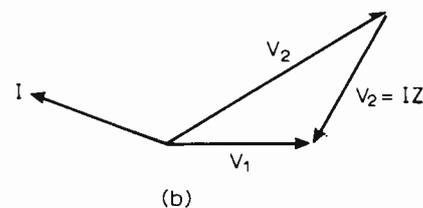


Fig. 9



(a)



(b)

Fig. 10

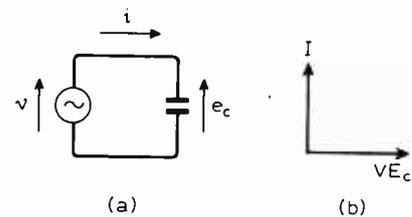


Fig. 11

The above example could represent an a.c. machine where V_2 is the generated e.m.f., V_1 the terminal voltage and Z the impedance. It could also represent a transmission line of impedance Z and voltages V_1 and V_2 at either end.

Capacitive circuits

In accordance with the principles already discussed, the circuit and phasor diagrams for capacitive circuits may be drawn as follows.

Simple capacitive circuit — Fig. 11(a)

When the current is positive, the voltage e_c will be increasing positively. Thus

$$i = +C de/dt.$$

In terms of a.c. phasors

$$I = +j \omega C E_c \text{ or } E_c = -I j / \omega C$$

Also

$$v - e_c = 0 \text{ or } v = e_c$$

The phasor diagram is shown at (b).

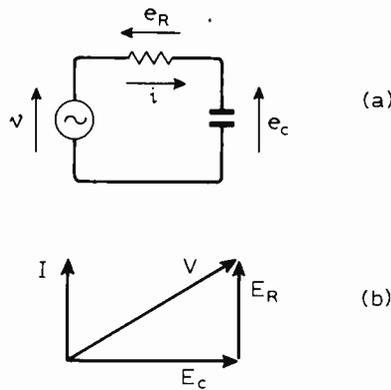


Fig. 12

RC circuit — Fig. 12(a)

We have

$$v - e_R - e_C = 0$$

$$v = e_R + e_C$$

The phasor diagram is shown at (b).

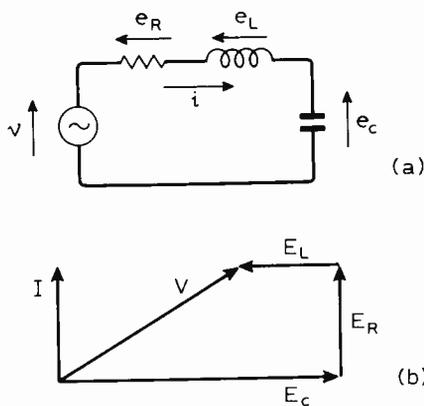


Fig. 13

RCL circuit — Fig. 13(a).

We have

$$v - e_R - e_L - e_C = 0$$

$$v = e_R + e_L + e_C$$

The phasor diagram is shown in (b).

Magnetically coupled circuits: transformers

For two magnetically-coupled circuits, i.e. a transformer. Assume as usual that an alternating magnetizing current and the flux produced by it are in phase i.e. they are proportional to each other (linear magnetic circuit assumed) and a positive current gives a positive flux. Thus the positive direction of flux linking a coil is that produced by a positive current. As self inductance is defined as the ratio of flux linkage to current, we are defining self inductance as a positive quantity.

Mutual inductance however (also defined as the ratio of flux linkage to current) will be taken as an algebraic quantity, positive or negative. Once positive flux directions are defined for both coils, one can say that if a positive current in one coil produces a flux in the negative direction in the other coil, then

the mutual inductance is negative and vice versa. Notice that we are giving more information than the commonly used dot notation, as the dot notation does not tell us the direction of flux produced by a current.

Now consider a transformer consisting of two co-axial coils wound in the same sense as shown in Fig. 14(a). For simplicity, assume unity turns ratio i.e. the two coils are identical. For the moment, neglect resistance and leakage flux i.e. assume maximum coupling. Then $L_1 = L_2 = M$.

Positive directions of voltage and current in the two windings have been selected as previously outlined; also the positive flux directions in both coils. As this is the same direction in both coils, the mutual inductance is positive. Notice that the power must be positive at one pair of terminals and negative at the other, i.e. input power at one pair of terminals and output power at the other. In electro-mechanical devices of course, it is possible to have positive power i.e. power input at both terminals. We then must obtain energy in some other form e.g. mechanical energy output.

Assume that the primary winding has a sinusoidal applied voltage V_1 and the secondary winding is open circuited. (The terms primary and secondary are purely arbitrary.) The instantaneous values of voltages e_1 and e_2

$$e_1 = +L_1 di_1/dt$$

$$e_2 = +M di_1/dt$$

Notice that e_1 and e_2 have the same sign if M is positive. This clearly must be so as a positive current in either coil produces flux in the same direction. The a.c. phasor equations are as follows

$$E_1 = +j\omega L_1 I_1$$

$$E_2 = +j\omega M I_1$$

E_1 and E_2 are in phase, as shown in Fig. 14(b)

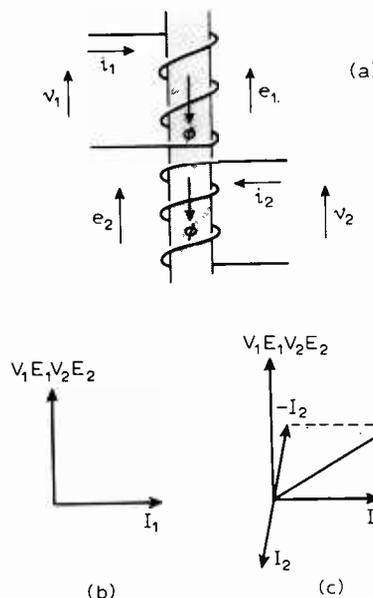


Fig. 14

Also for identical coils with maximum coupling $L_1 = L_2 = M$ and thus $E_1 = E_2$

Also

$$V_1 - E_1 = 0 \text{ and } V_2 - E_2 = 0$$

$$\text{Thus } V_1 = +E_1 \text{ and } V_2 = +E_2$$

If the secondary winding was wound in the opposite sense, making the mutual inductance negative, then e_1 and e_2 would be of opposite sign, and E_1 and E_2 would be in anti-phase. However, the previous conditions can be restored merely by reversing the positive directions of current and voltages in the secondary winding.

Reversing the positive current direction will restore M to a positive value and thus reverse the sign of e_2 . But the reversed value of e_2 acts in the opposite direction due to the reversal of the positive e_2 direction. Thus the answers obtained are still correct.

Now consider the transformer to be carrying a load current I_2 due to the connection of an impedance across the secondary terminals. The current I_2 is shown in the phasor diagram of Fig. 14(c). V_2 and I_2 represent a negative power i.e. output power. There will be an additional current $-I_2$ in the primary circuit whose flux cancels that of I_2 i.e. always of opposite sign and of equal magnitude i.e. in phase opposition, Fig. 14(c). Otherwise the total flux and also E_1 would alter. This cannot be so since E_1 is equal in magnitude to V_1 , the r.m.s. value of which is assumed to be constant. The total primary current is given by

$$I_1 = I_{1oc} + (-I_2)$$

where I_{1oc} is the value of I_1 with the secondary open circuited, Fig. 14(b).

E_1 and E_2 are now given by

$$E_1 = +j\omega L_1 I_1 + j\omega M I_2$$

$$E_2 = +j\omega L_2 I_2 + j\omega M I_1$$

As $L_1 = L_2 = M$ for identical coils and maximum coupling, then

$$E_1 = E_2 = +j\omega L(I_1 + I_2)$$

As $I_1 + I_2 = I_{1oc}$ from the equation above, we see that the values of E_1 and E_2 are unchanged from the open circuited condition, as already stated.

Now include winding resistances R_1 and R_2 and leakage inductances l_1 and l_2 . The diagrams will now be as shown in Fig. 15 (a) (b). The two circuit equations will be

$$V_1 - E_{R1} - E_{l1} - E_1 = 0$$

$$V_2 - E_{R2} - E_{l2} - E_2 = 0$$

where

$$E_{R1} = +R_1 I_1 \text{ and } E_{R2} = +R_2 I_2$$

$$E_{l1} = +j\omega l_1 I_1 \text{ and } E_{l2} = +j\omega l_2 I_2$$

Thus

$$V_1 = I_1 R_1 + j\omega l_1 I_1 + E_1$$

$$V_2 = I_2 R_2 + j\omega l_2 I_2 + E_2$$

$$\text{Also } E_1 = E_2 = +j\omega L(I_1 + I_2)$$

Iron losses

Iron losses can be taken into account in the usual way by inserting a suitable resistance across the voltage e_1 . The phasor representing the no load current is then at an angle less than 90° to the voltage e_1 .

Alternative transformer diagrams

The diagrams shown in Fig. 16 show alternative circuit and phasor diagrams as they are often drawn for the transformer. The diagrams also show the conventions which have been assumed, although these are usually not stated at all.

The equations are

$$V_1 = (-E_1) + I_1 R_1 + j\omega l_1 I_1$$

$$E_2 = V_2 + I_2 R_2 + j\omega l_2 I_2$$

Notice the following points.

● It is assumed that if winding impedances are neglected, the applied voltage and induced e.m.f. in the primary are equal and opposite. In the secondary, induced e.m.f. and terminal voltage are of the same sign. Notice the inconsistency.

● Positive electrical power in the primary is power input; positive power in the secondary is power output.

● As the secondary current i_2 and the compensating primary current are equal and opposite, one concludes that a positive current in both windings give fluxes in the same direction. However, this convention is rarely, if ever, stated. As mentioned previously, if it is stated and positive currents are clearly defined, there is no need for the dot convention.

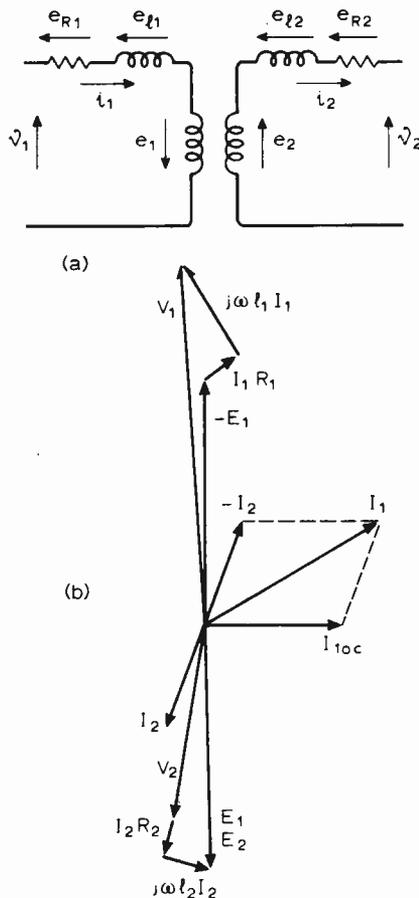


Fig. 16

General coupled circuit theory

The previous discussion has been concerned with the so-called "power transformer theory." Now consider the general coupled circuit theory using self and mutual inductances.

Refer again to the coupled coil system shown in Fig. 14 (a). The two circuit equations can be written

$$v_1 = i_1 R_1 + L_1 \frac{di_1}{dt} + M \frac{di_2}{dt}$$

$$v_2 = i_2 R_2 + L_2 \frac{di_2}{dt} + M \frac{di_1}{dt}$$

For sinusoidal operation, the equations become

$$V_1 = I_1 R_1 + j\omega L_1 I_1 + j\omega M I_2 \quad (1)$$

$$V_2 = I_2 R_2 + j\omega L_2 I_2 + j\omega M I_1 \quad (2)$$

Let V_1 be an applied voltage and let I_2 be obtained by connecting an impedance $Z = R + j\omega L$ across the winding.

$$V_2 = Z I_2 \quad (3)$$

Notice that with the positive directions defined as in Fig. 14 (a) a positive voltage applied to Z would give a negative current. Hence the minus sign in equation (3).

Equations (1), (2) and (3) can be solved to give

$$\frac{V_1}{I_1} = R_1 + \frac{\omega^2 M^2}{|Z_S|^2} \cdot R_S + j\omega \left(L_1 - \frac{\omega^2 M^2}{|Z_S|^2} \cdot L_S \right) \quad (4)$$

where

$$R_S = R + R_2, L_S = L + L_2, Z_S = R_S + j\omega L_S$$

Also equations (2) and (3) give

$$I_1 = -I_2 \frac{(R_2 + j\omega L_2 + Z)}{j\omega M} \quad (5)$$

From equation (5) the phasor for I_1 leads I_2 by an angle between 90° and 180° . From equations (1), (2), (4) and (5) we can sketch the phasor diagram as shown in Fig. 17. The phasors for V_1, V_2, I_1 and I_2 in Fig. 17 would of course be identical to those in Fig. 15.

Networks with more than one mesh

Fig. 18 shows a more complex network with more than one mesh (three in this case.) The positive directions of the three mesh currents are indicated. Also, the positive directions of the branch currents in the inductances have been indicated to obtain mutual inductance voltage terms correctly. The voltage equation for mesh three is

$$V_1 = R_1 I_3 + j\omega L_1 (I_3 - I_1) + j\omega M_{13} (I_2 - I_1) + j\omega M_{12} (I_3 - I_2) + j\omega L_2 (I_3 - I_2) + j\omega M_{23} (I_2 - I_1) + j\omega M_{21} (I_3 - I_1) - j/C_3$$

The other equations will be obtained similarly.

Before the problem can be completed, the sign of the various mutual inductances must be specified. For example, if a current in the indicated positive direction in L_3 produces a positive flux in L_1 i.e. a flux in the same direction as a current in the indicated positive direction in L_1 would produce, then M_{13} is positive. This information could be given by means of the dot convention, but it should be noted that when there are more than two coils, it may be necessary to use different types of dots for different pairs of coils.

Electro-mechanical device

In the simple electro-mechanical device shown in Fig. 19, mechanical energy is involved by the movement of the

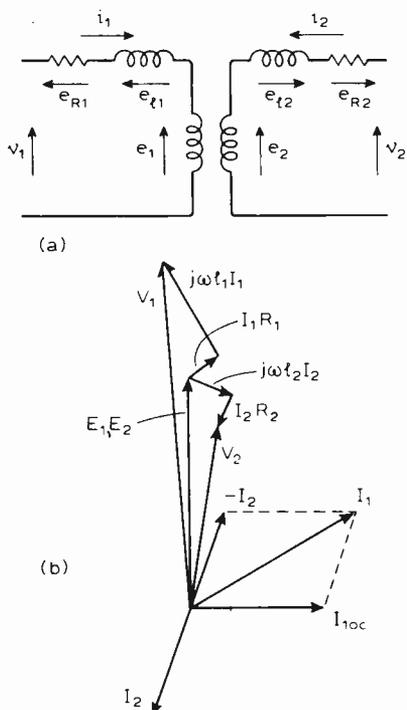


Fig. 15

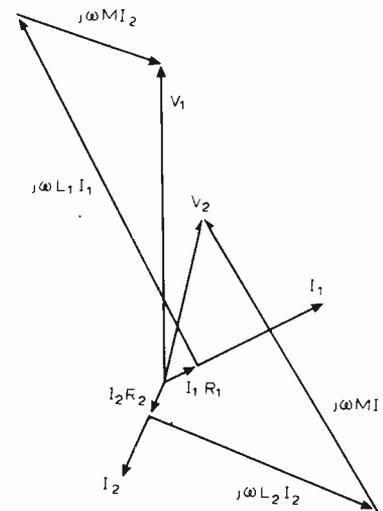


Fig. 17

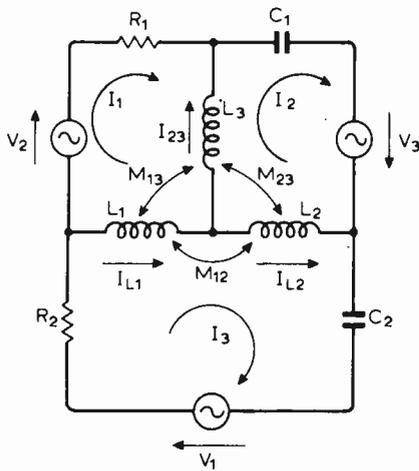


Fig. 18

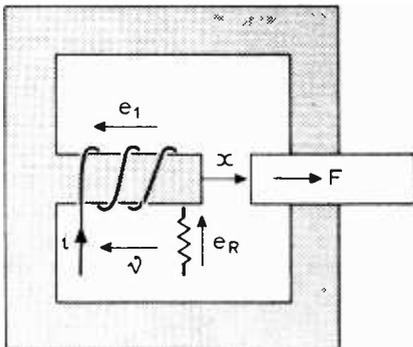


Fig. 19

plunger under the action of the electromechanical force on it. The assumed positive directions of the electromagnetic force F and the distance of the plunger from a fixed point x are shown in Fig. 19. Mechanical energy will be produced by the device if the force F moves in the direction of the force i.e. if F and dx/dt are both positive or both negative.

The mechanical power is

$$P_m = Fdx/d\dot{x}$$

Thus a positive value of P_m signifies mechanical power output.

The electrical power input to the coil is

$$P_e = v i = i \left(\frac{d}{dt}(Li) + Ri \right) = i^2 R + iL \frac{di}{dt} + i^2 \frac{dL}{dt}$$

The rate of increase of stored energy in the field is

$$P_f = \frac{d}{dt} (\frac{1}{2} Li^2) = Li \frac{di}{dt} + \frac{1}{2} i^2 \frac{dL}{dt}$$

Now if P_h is the energy converted to heat, i.e. $i^2 R$, the energy equation is as follows

$$P_e = P_h + P_f + P_m$$

$$i^2 R + iL \frac{di}{dt} + i^2 \frac{dL}{dt} =$$

$$Li \frac{di}{dt} + \frac{1}{2} i^2 \frac{dL}{dt} + F \frac{dx}{dt} + i^2 R$$

$$F \frac{dx}{dt} = \frac{1}{2} i^2 \frac{dL}{dt}$$

$$F = \frac{1}{2} i^2 \frac{dL}{dx}$$

This is a familiar answer, but in which direction is the force? Clearly, from the diagram, dL/dx is negative, i.e. for increase in x , L decreases. Also i^2 must be positive. Therefore F must be negative, i.e. in the direction such as to close the gap.

Books Received

Control Technology — pupils' assignments and follow-up sheets are two tutorial books aimed at pupils who are taking science subjects. The first publication deals with mechanics and electronics projects, the second book is a progression from the first with the same subjects being dealt with in more detail. Price £1.80 each. Pp.108 and 150. The English Universities Press Ltd, Saint Paul's House, Warwick Lane, London EC4V 4AH.

Small Appliance Repair Guide Vol. 2 by Leo G. Sands, **4 Channel Stereo from Source to Sound** by Ken Sessions, **Practical Circuit Design for the Experimenter** by Don Tuite, and **Basic Digital Electronics** by Ray Ryan. Tab Books, Blue Ridge Summit, Pa.17214, U.S.A.

Wideband Voltage Amplifiers by C. W. Davidson, is a soft-back handbook which explains the design procedure of amplifiers operating from d.c. to tens of MHz. Basic principles and circuits, feedback, input/output stages, and high gain amplifiers are dealt with in six chapters, a concluding section covers applications of high gain d.c. amplifiers. Price £1.95, Pp.112. Intertext Publishing Ltd, 158 Buckingham Palace Road, London SW1W 9TR.

Electrical Installations and Regulations by Michael Neidle is intended to assist students studying for the City & Guilds electrical installation work, (No. 235) course B certificate. The Macmillan Press Ltd, 4 Little Essex Street, London, W.C.2.

A Review of the Principles of Electrical and Electronic Engineering Vol.4 — **Microwaves, Communications and Radar** edited by L. Salymer. Six chapters by four authors discuss microwaves, aeriels, communication theory, noise, communication systems and radar. The book is suitable for students or those interested in the above topics and contains formulae and diagrams where applicable. Price £2.50. Pp.190. Chapman & Hall Ltd, 11 New Fetter Lane, London EC4P 4EE.

Summary

If the positive directions of all quantities are clearly defined, writing and solving the equations of any electrical network or electro-mechanical device is straightforward and no confusion should arise. The procedure is summarized below.

- Indicate chosen positive directions of currents.
- Indicate positive voltage directions by arrows (heads positive). The branch of the network will be as shown in Fig. 20 (a) or (b).
- note the meaning of positive power. In Fig. 20 (a) positive power is dissipated power. In Fig. 20 (b) positive power is generated power.
- If a branch of the network contains an impedance Z , we have

$$V_z = +ZI \text{ Fig. 20 (a)}$$

$$\text{or } V_z = -ZI, \text{ Fig. 20 (b)}$$

where Z is R or $j\omega L$ or $-j/\omega C$ for a.c. sinusoidal conditions.

- For circuits with magnetically coupled coils, determine the algebraic sign of all mutual inductances with reference to the chosen positive current directions in the inductances.

$v_z' = +ZI'$ using the system of Fig. 20 (a)

or $V_z' = -ZI'$ using the system of Fig. 20 (b)

where $Z = \omega M$ and M is the mutual inductance between the impedance across which V_z' exists and an impedance in another branch carrying a current I' .

- The electrical network equations can now be written.

- For mechanical devices, define positive directions of force, position, speed, etc and note the significance of positive mechanical power, i.e. input power to the device or output power.

The results of calculations will now give the correct signs and thus directions of all quantities.

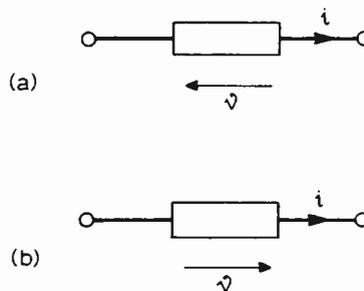


Fig. 20

High-stability VHF receiver

The 1990R Series of Eddystone General Purpose VHF Receivers provides reception facilities for AM, FM, CW and pulse transmissions.
Model 1990R/1 covers the band 25-235MHz.
Model 1990R/2 has additional ranges extending the coverage to 25-500MHz.

All 1990R receivers are equipped for high-stability working with either a synchroniser (illustrated) or a 10-channel crystal facility. Provision is also made for operation with externally derived oscillator signals as an alternative to continuous tuning with the free-running local oscillator.



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A clear view of the band

Eddystone EP961 MkII Panoramic Display Units provide visual monitoring of all signals in a selected band.
MkII-A is tunable from 50kHz to 800kHz, matching the IFs used in MF and HF communication receivers.

MkII-B covers 500kHz to 36.5MHz. It is ideal for use with VHF and UHF receivers for monitoring FM broadcasts and communication transmissions, and its usefulness extends into the laboratory field.

Both versions can be used with direct aerial input in many applications.



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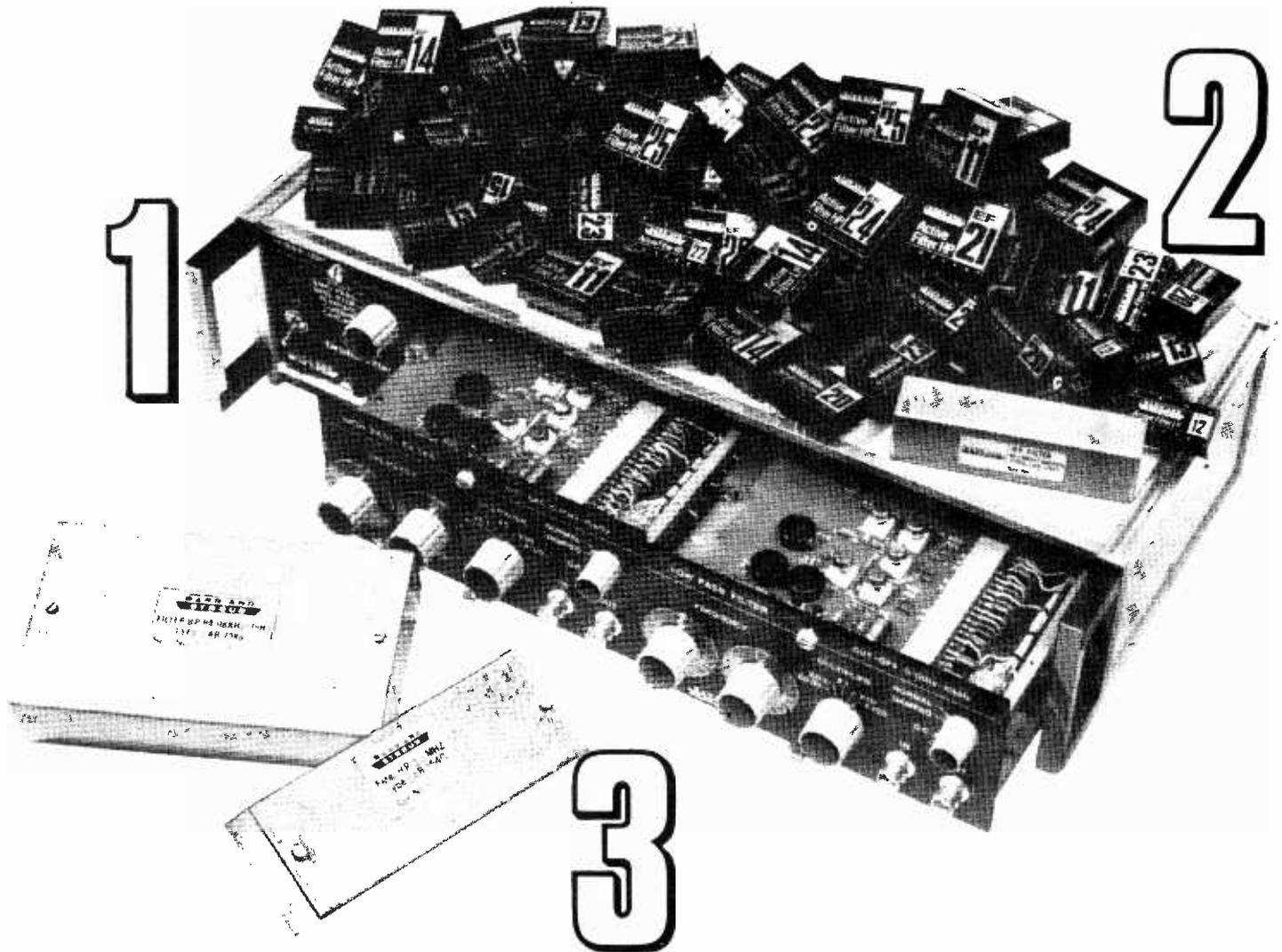
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These are compact, solid state, encapsulated units providing basic filter functions to be customer set for cut-off frequency and characteristic. The present range contains Low Pass and High Pass types with cut-off frequency coverage from 1.0Hz to 30kHz in overlapping ranges, with attenuation rates up to 24dB/octave/module. Universal modules specifically for Band Pass and Band Stop operation are part of the range.

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Last Intelsat IV launch

The seventh link in the current chain of global communications satellites now ringing the Earth was launched earlier this year from Cape Canaveral, USA. The 3,000-pound satellite was launched towards a 22,300-mile-high synchronous orbit to become the second such spacecraft over the Indian Ocean. Operation of the satellite segment of the Intelsat-owned system is conducted by the Communications Satellite Corporation. Comsat said that each satellite has an average use of 4,000 telephone circuits or a capability of providing 12 television channels, or various combinations of telephone, television and data transmissions. The current Intelsats are the fourth and largest generation of communications satellites since Early Bird (Intelsat I) joined Western Europe and the United States with telephone and television service ten years ago. Since 1965 trans-oceanic telephone traffic has grown from an estimated three million calls a year to more than 50 million in 1974, a steady growth rate of 20 per cent annually. To handle the predicted growth, Hughes, under contract to Intelsat, is building a new generation of satellites called Intelsat IVA. Six spacecraft are under construction. The first in the series is scheduled to be launched from Cape Canaveral this summer for service over the Atlantic Ocean.

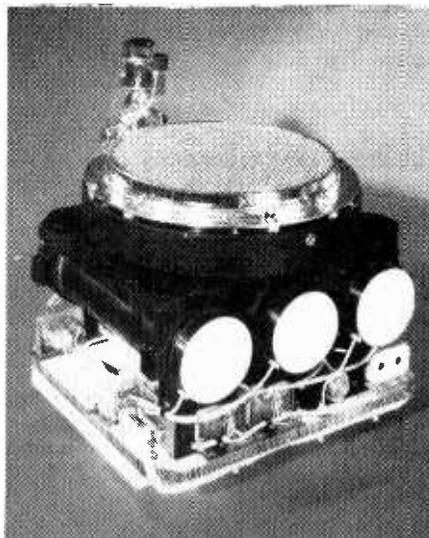
Communications satellite moves to India

The largest and most powerful communications satellite ever built was moved a third of the way around the globe early in June to place it in a position for an instructional television experiment involving several thousand villages in India. NASA's Applications Technology Satellite-6 was launched into a geosynchronous orbit on May 30, 1974 and has since been used in the

experimental tracking of spacecraft, ships and aircraft and also in the use of satellite television for education and even medical diagnosis. The satellite provided one communications link during the US-Soviet Apollo Soyuz test project. After ATS-6 arrived at its new station about July 1, it was checked out prior to use in mid-July to track the Apollo and Soyuz spacecraft and to relay television and data from the orbiting spacecraft to Earth. This was the first time that a satellite had been used to relay television from a manned spacecraft to the planet's surface.

The satellite move was designed to bring it within range of India for the Satellite Instructional Television Experiment, scheduled to be formally

Rotation modulation collimator, Experiment A, on board Ariel 5, the Science Research Council's X-ray astronomy satellite, was an experiment to accurately determine the positions of cosmic X-ray sources. The accuracy in determining the position of a source is limited by counting statistics and accuracy in the knowledge of the satellite's pointing position, which can be determined by the experiment itself if an already well determined source is in the field.



inaugurated on August 1. The experiment is a joint effort by NASA and the Indian Space Research Organization. In this experiment, the spacecraft will be used by the Indian Government to relay daily TV programmes to 5,000 villages and cities in seven states through India. About half the villages will be equipped with TV sets augmented by converters and small antennas to receive the television signal after it is rebroadcast from a ground terminal in the area. Following the year-long experiment in India, the satellite will be returned to a position in range of the USA for further experimental use.

ASTP's ranging system

The v.h.f. ranging system for the Apollo Soyuz Test Project used the Command Module and Soyuz v.h.f. link to provide a ranging capability of 200 nautical miles with ± 250 feet accuracy. The system consists of a Lunar Module v.h.f. set, a Command module v.h.f. set, a Lunar Module Ranging Tone Transfer Assembly (RTTA) and Command Module Digital Ranging Generator (DRG) from the Apollo programme. The Lunar Module v.h.f. sets and Ranging Tone Transfer Assemblies were installed in the Soyuz.

European Space Days

Approximately 200 people from 30 countries including Algeria, Brazil, Canada, India, Iran and Nigeria participated in the European Space Days, May 27-29, organized at the European Space Research and Technology Centre, Noordwijk, Netherlands. The aim was to demonstrate Europe's space capability to an invited audience representing telecommunications authorities and other users of space systems.

Speaking at the presentation of European space programmes, the Secretary General of the International Telecommunications Union, Mr M. Mili, said that the ITU welcomed and encouraged Europe's firm intention to play an important role in the peaceful uses of outer space. He believed that the programmes of the new European Space Agency offered a particularly well-chosen range of satellites for research and for applications of unquestionable value. The Secretary General of Eurospace, Mr Y. Demerliac, recalled that since 1962 Europe had developed 43 satellites, including four for telecommunications and/or television. European industry had built 36 of the 104 ground stations or antennas of the Intelsat network, and had sold more than two-thirds of these to non-European countries.

World of Amateur Radio

Less television interference

The latest television and radio interference complaints report from the Home Office Radio Regulatory Department — covering 1974 — shows a marked decline in the total of all complaints (—21.7% to 48,371) and also in those ascribed to amateur radio transmitters (—26.6% to 886). This is the first time for many years that the number of complaints traced to amateur operation has fallen below a thousand: l.w./m.w., 75; Band 1, 140; Band 2, 71; Band 3, 108; Bands 4-5, 480; mobile services, 12. The biggest decline is in complaints to v.h.f. television which are down by over a half, with u.h.f. complaints holding fairly steady. Part of the credit must go to the television industry which at last seems to be making efforts to design u.h.f. domestic receivers reasonably immune to strong local signals. Indeed, if only television sets did not radiate so many harmonics from line timebases and switched-mode power supplies it seems that real progress could be made in bringing viewers and amateur radio into a happy relationship.

SOE's suitcase sets

Still to be heard on the amateur bands are numbers of the B2 suitcase sets produced during the war for the Special Operations Executive (later part of Special Forces) the British "cloak and dagger" organisation set up in July 1940. SOE's appointed task, until it disbanded in 1946, was to co-ordinate and initiate subversive and sabotage activities against the enemy and its Signals Directorate was responsible for providing clandestine radio links into Europe and elsewhere, independently of the secret Intelligence radio circuits.

Many of the SOE signals people were, or later became, radio amateurs, including some of the base station operators who were often girls recruited into FANY.

Sir Colin Mc V. Gubbins has claimed that the most valuable link in secret

operations were the agents with h.f. c.w. sets — "without these links we would have been groping in the dark." Tragically such communications were extremely vulnerable to enemy interception and "funkspiel" radio games, such as the German North Pole operation in Holland, and much of SOE's later signals work was in developing techniques to speed up communications to counter the highly organised D/F teams. They also produced many methods of charging batteries in the field including the "beach chair" pedal generator, bicycle generators, wind generators and even portable steam-driven generators and thermo-couple chargers.

At the recent 30th anniversary dinner of the Special Forces Club, at which Prince Charles was guest of honour, a display was staged on the theme of communications. Codes and signal plans were provided by Leo Marks; 450MHz S-phone equipment, originally conceived in 1941 by the late Bert Lane, was shown by Flt Lt Charles Bovill; and suitcase sets and MCR1 miniature communications receivers by Major John Brown, G3EUR, who was largely responsible for the development of the A2, A3, B1, B2, B3 and MCR1 equipments.

Sun shines on 28MHz

Mid-summer in a sunspot minimum year is not the most likely time for transatlantic 28MHz openings: the exception that proves the rule that you can never be sure was the evening of July 4. K3NPV was so surprised at what he found on "ten" that he changed to 21MHz and sent out a "QST" message: "28MHz is wide open to Europe," as indeed also was 21MHz. By that day Mike Matthews, G3JFF of Portsmouth, had already worked 28 countries on 28MHz this year and even when the F layer is reluctant to bounce back 28MHz signals, Sporadic E can provide useful contacts. Early July was also a good period for v.h.f. propagation with the Gibraltar beacon reported heard in London on July 2 and Danish stations worked through the London GB3LO repeater on July 5.

When portable is fixed

My dictionary defines "portable" as "movable (article), convenient for carrying" but it seems the Home Office is having semantic problems over this: suddenly it has become hot and bothered about amateurs using hand-held equipment "while walking." In a curious decision it has pronounced that the operation of such equipment is at present covered neither by the normal licence (which permits "portable" operation) nor by the amateur (sound mobile) licence; walking use of equip-

ment is therefore being sanctioned by a special letter of authorisation to the mobile licence. This means amateurs will need the extra mobile licence to exercise this facility: presumably if you stand still you are working "portable" but if you move while transmitting you become a "pedestrian mobile." It seems a little like the old regulation that allowed the young ladies at the Windmill Theatre to appear in the altogether only so long as they maintained a rigid pose, and presumably the Home Office is now tackling the problem of how many amateurs can transmit while dancing on a pin head.

In brief

Honorary secretary of the Radio Amateur Invalid and Bedfast Club, is now Mrs Rita Shepherd, G3NOB, 59 Pantain Road, Loughborough, Leics, LE11 3LZ.... With five 144MHz repeater stations now operational in the UK the FM Group London reports that additional stations are being planned for Martlesham, Suffolk; Barnsley, Yorkshire; Black Hill near Glasgow; Buxton, Derbyshire; Luton, Beds (432MHz); Bacton, Norfolk; Cambridge (432MHz); Newquay, Cornwall; Aberdeen; Birmingham; and Carmel, Dyfed. The Carmel repeater hopes to use the callsign GB3WW which some readers recall was used from the *Wireless World* offices at Dorset House in connection with the journal's 60th anniversary during April 1971.... At Kingston-upon-Thames Magistrates Court an amateur recently pleaded guilty to causing interference and using a station for wireless telegraphy without a licence in connection with interference to the London repeater station GB3LO.... September mobile rallies include those at Peterborough (Walton School) on September 21; Castle Grounds, Antrim on September 21; and Netteswell School, Harlow on September 28.... Paid up membership of the British Amateur Radio Teleprinter Group is almost 350 of which about 230 hold British callsigns and some 100 British r.t.t.y. stations are active on 144 MHz.... An amateur 625-line PAL colour television station with an effective radiated power of 12kW (vision) and 600 watts (sound) is being operated daily in Sydney, Australia by Vic Barker, VK2ZVV/T and Ian MacKenzie, VK2ZIM/T with a vision frequency of 442.3MHz and a 64-element phased array aerial.... The GB3SN repeater at Four Marks, Hampshire is now fully operational on 145.125MHz input, 145.725MHz output. Coverage extends from Devon to Kent and north into Wales and Cambridge. The station is run by the UK FM Group (Southern) and the transmitter has an e.r.p. of 70 watts from aerials 250 metres above sea level (access tone 1750Hz for 0.5s followed by callsign).

PAT HAWKER, G3VA

New Products

Temperature protection unit

A temperature protection unit designed for use with positive temperature coefficient thermistors (to BS4999) and intended for the protection of electric motors, has been introduced by ITT. The thermistors are embedded in the winding of motors and electrically connected to the ZK1 unit which is mains powered and has a switching capability up to 2.2kVA at 240V a.c. ITT Components Group Europe, Thermistor Division, Stephen Street, Taunton, Somerset.

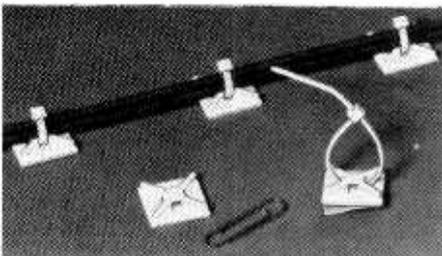
WW 310 for further details

Adhesive cable clips

A small adhesive cable clip measuring about two cm square will secure a bundle of cables up to five cm diameter and support up to 55g. To fix the clip a backing paper is removed and the



WW310



WW302

mount is pressed onto the required surface which must be clean, dry and smooth. Panduit Ltd, Sittingbourne Industrial Park, Unit 22a, Crown Quay Lane, Sittingbourne, Kent.

WW 302 for further details.

Desoldering gun

The Ersa Vac 40 is a de-soldering gun that operates by melting the solder and "sucking up" the molten metal via a regulated vacuum. A separate mains-powered power/vacuum supply is provided and the total power consumption is around 50W. Greenwood Electronics, Portman Road, Reading RG3 1NE.

WW 303 for further details

Ten-turn potentiometer

A ten-turn potentiometer, model 3541, uses a new composite called Hybriton which combines a wirewound and conductive plastics element. The makers say this provides a potentiometer with long life, infinite resolution, and a low temperature coefficient. The component is available with resistances from 1 to 100k Ω , has a linearity within $\pm 0.25\%$, and a resistance tolerance of $\pm 10\%$. Bourns (Trimpot) Ltd, Hodford House, 17 High Street, Hounslow, Middx TW3 1TE.

WW 305 for further details

Charge-coupled image sensor

The SID51232 from RCA is a 512 \times 320 element silicon image sensor which can replace conventional TV camera tubes.

This sensor is a self-scanned device based on c.c.d. technology and is claimed to be the first solid-state image sensor to generate standard 525-line video, compatible, with existing TV monitors and accessories. The sensor is supplied in a hermetically-sealed, 24-connexion ceramic package containing an optical glass window. RCA Electro-Optics & Devices Division, Lincoln Way, Sunbury-on-Thames, Middx. (See August issue, p.362).

WW 317 for further details

High-voltage resistors

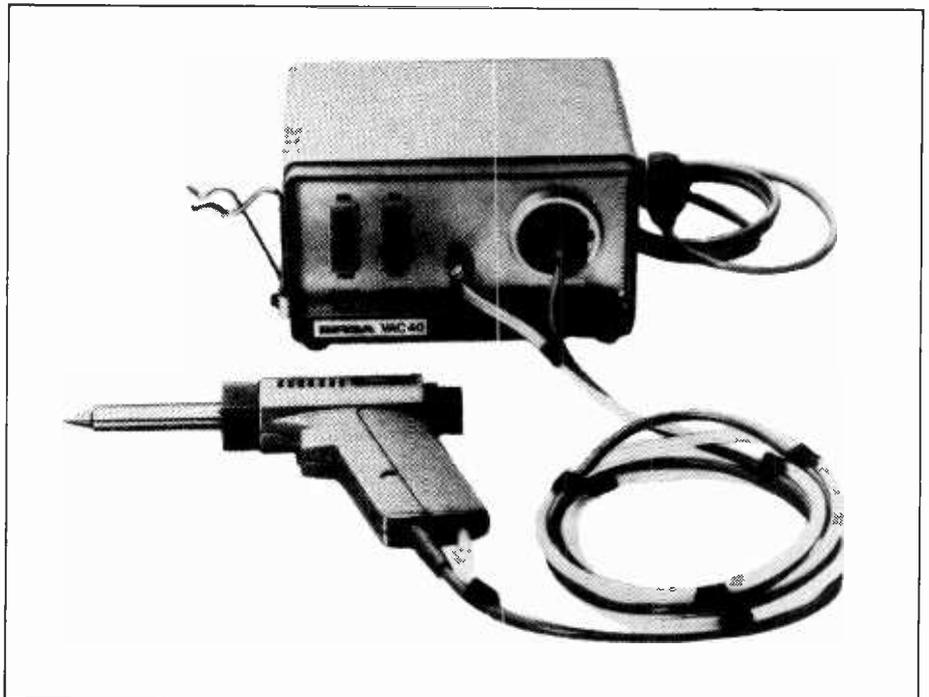
Mullard has introduced a new range of metal-glaze resistors which will withstand 10kV d.c. at power levels up to 1W. Resistance values extend from 1 to 68M Ω with a tolerance limit of $\pm 5\%$, and resistance stability of typically less than 1% change after 1000 hours at 0.5W in an ambient temperature of 70°C. Mullard Ltd, Mullard House, Torrington Place, London WC1.

WW 304 for further details

F.m. - receiver filters

The Toyocom TQF 2599/3079 and QN 0071 are claimed to be the lowest price crystal-filters available for broadcast-band stereo f.m. receivers. The filters are designed to be used in pairs — one being separated from the other by an isolating amplifier stage. In this configuration the filters provide a 3dB bandwidth of 120kHz either side of the 10.7 MHz i.f., with the -70dB point at around $\pm 300\text{kHz}$. Walmore Electronics Ltd, 11 Betterton Street, London WC2H 9BS.

WW 309 for further details



WW303

Micromotors

The Escap 23D series of motors comprises the 23D21-216 and 23D21-213 which offer a mechanical time constant of 10ms; moments of inertia of 5.80 and $4.45 \times 10^{-7} \text{ kgm}^2$ respectively; starting torques of 295 and $245 \times 10^{-4} \text{ Nm}$ respectively. The motors, which are claimed to have an efficiency of around 80%, operate from 12 and 15V, have outputs of 3.8 and 3.4W with no-load speeds of 4900 and 5250 rev/min. Portescap (UK) Ltd, 204 Elgar Road, Reading, RG2 0DD.

WW 301 for further details

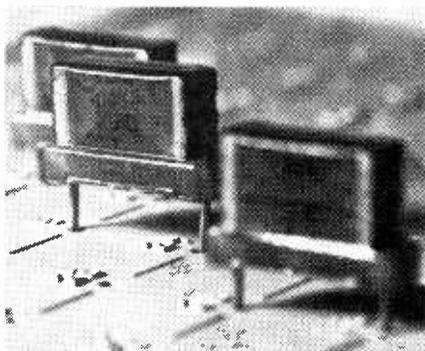
Capacitors

A range of miniature metalized plastic-film capacitors designated type MKM are manufactured by Siemens. Components are cut from a "mother" capacitor of known value, in this way a uniformity of electrical characteristics is achieved. Ratings are 0.01 to $0.68 \mu\text{F}$ at 100V d.c., and 0.1, 0.15 and $0.22 \mu\text{F}$ at 250V d.c. LST Electronic Components Ltd., Victoria Road, Chelmsford, Essex.

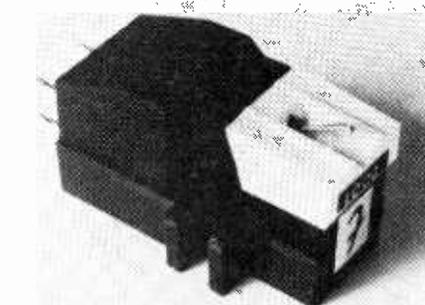
WW 311 for further details



WW301



WW311



WW314

Magnetic cartridge

Condor Electronics are now marketing the Tenorel T2001 magnetic cartridge. This is a low priced unit (£5.45 excluding VAT) which is claimed to equal the performance of cartridges such as the Shure M75/B, the Philips GP400, and the Goldring G820. Manufacturers specifications for the device are: frequency response 15Hz to 23kHz; separation more than 25dB at 1kHz; output 5.5mV at 1kHz and 5cm/s; tip mass 1mg; and a playing weight of between 1½ and 3 grammes. Condor Electronics Ltd, 100 Coombe Lane, London SW20 0AY.

WW 314 for further details

Logic-state analyzers

The model 1600A is a self-contained analyzer incorporating a c.r.t. which can display a 16-channel sequence of data in word form using the 1 and 0 format. A model 1607A does not have a c.r.t., but provides both analogue and digital outputs. These analogue outputs can be used to convert most oscilloscopes, with d.c. coupled X, Y and Z inputs, into an analyzer display. If the two models are combined a 32-channel

display can be achieved with both units capable of operating at clock speeds up to 20MHz. If used separately, the instruments may operate with different clock rates and the two displays can be compared using an exclusive-OR comparison. Hewlett-Packard Ltd, King Street Lane, Winnersh, Wokingham, Berks RG11 5AR.

WW 315 for further details

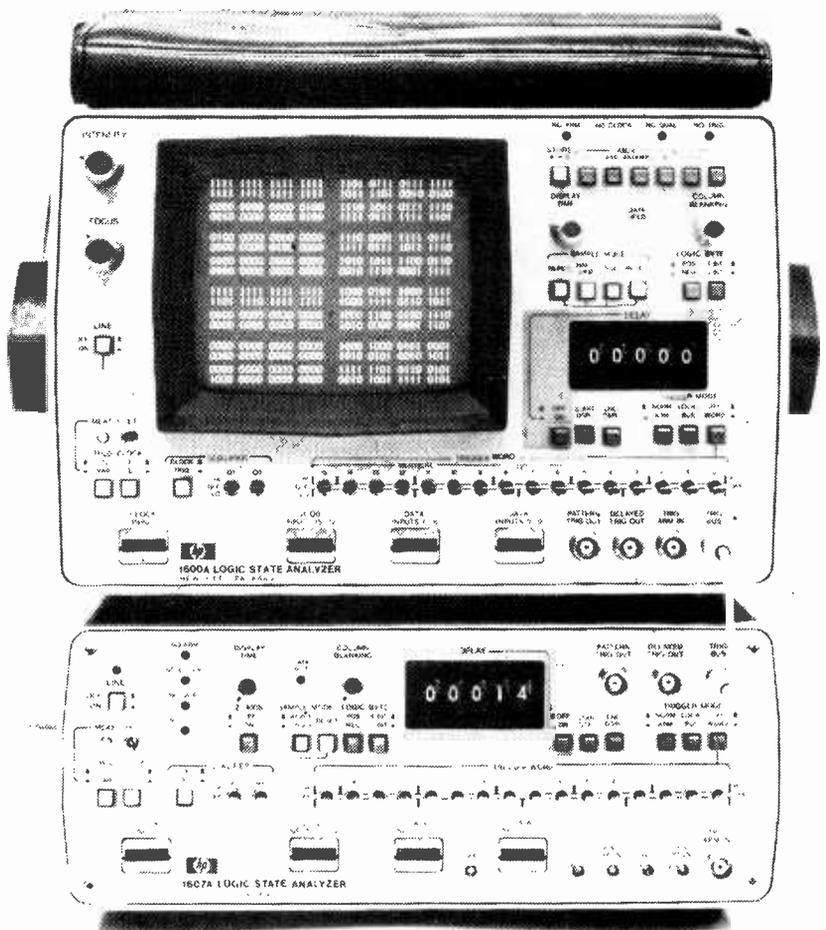
F.m. intercom

A frequency-modulated intercom system provides speech communication by transmitting through the 240V a.c. mains supply lines. Once the stations are located and connected to the supply the system will transmit up to distances of around 1 km provided that both stations are plugged into the same power line. Hadley Sales Services, 112 Gilbert Road, Smethwick, Warley, West Midlands B66 4PZ.

WW 313 for further details

Transient generator

The model 510 is a transient generator which produces signals that comply with the I.E.E.E. standards for surge



WW315

capability tests. The instrument provides bursts of damped sine waves with amplitudes adjustable up to 2.5kV and 50% decay times around 6 μ s. The bursts can be single-shot, free running, or synchronized to mains frequency. Nominal frequency of the sinewave oscillator is 1.25MHz and the output is available at selectable impedances of 100, 150, 600 or 1200 ohms. Euro Electronic Instruments Ltd, 27 Camden Road, London NW1 1YE.

WW 306 for further details

Circular-chart recorder

A circular-chart recorder that offers radial recording of information has a 10in diameter chart and can be equipped with up to three pens. The recorder is fitted with a stepping-motor servo system which is claimed to provide high accuracy without overshoot. Rotational speeds of the instrument are one revolution every 8, 12 or 24 hours. Full scale response is 0.8s from any input between 50 and 300mV and any direct current from 100 μ A to 100mA or alternating current from 250 μ A to 500mA, selectable by plug-in cards. These

ranges may be extended by means of external shunts/current transformers. Boyle Industrial Gauging Systems Ltd, Burch Road, Northfleet, Kent DA11 9NE.

WW 307 for further details

Low-current scanner

Keithley Instruments have introduced a low-current scanner designed for switching currents from picoamperes to tens of milliamperes. The unit provides ten channels of single-pole switching; channels not connected to the output are grounded to complete current paths. Front panel controls of the scanner permit either manual channel selection or automatic sequential scanning. The automatic scan rate is variable from 10 channels per second to 1 channel per ten second. Keithley Instruments Ltd, 1 Boulton Road, Reading, Berks.

WW 316 for further details

Batteries

The range of batteries known as Wonder Top are individually stamped with the stated shelf-life which is up to

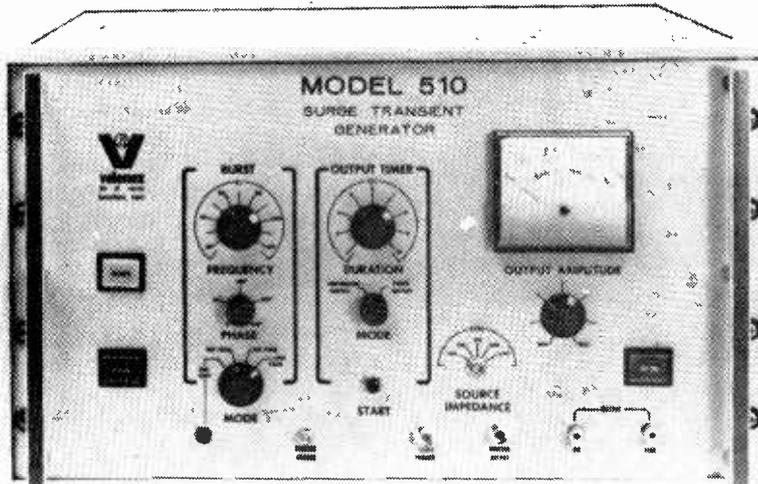
two years at 19°C. The batteries have a plastic cap and a protective cover over the positive terminal. The cap can be unscrewed for testing and overscrewed, breaking off the cover, for installation. The batteries, which are suitable for instrumentation, are priced at between 9 and 25p depending on quantity and type, and are available from West Hyde Developments Ltd, Ryefield Crescent, Northwood, Middx HA6 1NN.

WW 308 for further details

Microphone calibrator

A high pressure microphone calibrator, type 4221, consists of a pressure exciter with piston, a high-pressure coupler and a low-frequency coupler. It provides a frequency range up to 1kHz and down to 0.001Hz with the two couplers respectively. Due to a high ratio between the mass of the exciter body and the mass of the moving element (150:1) and the small piston amplitude the vibration level at the microphone is small which, say the manufacturers, can be an important feature. B & K Laboratories Ltd, Cross Lances Road, Hounslow, Middlesex.

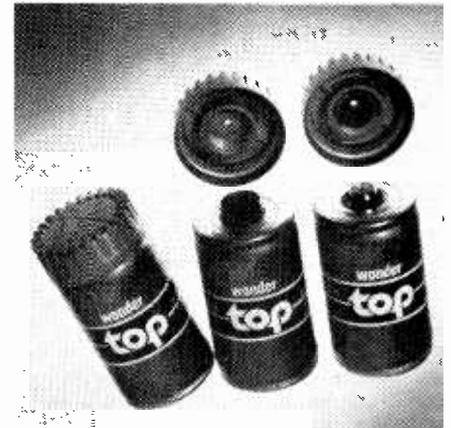
WW 312 for further details



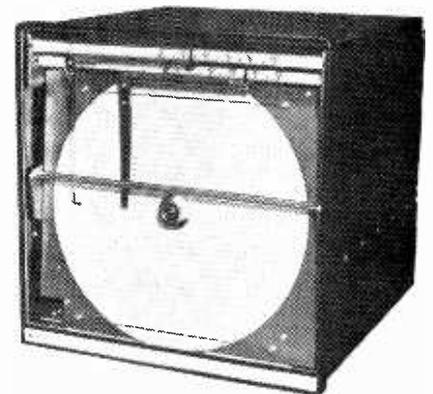
WW306



WW316



WW308



WW307

Solid State Devices

Names of suppliers of devices in this section are given in abbreviation after each entry and in full at the end of the section.

Vertical deflection i.c.

All the functions of a colour TV receiver vertical deflection system — oscillator, voltage ramp generator, pre-amplifier and power amplifier — are available from the TDA 1270 i.c. The chip is housed in a 12-pin dual in-line package and operates from supplies up to 40V.

WW351 for further details SGS ATES

Microcomputer

A p-channel single m.o.s.-chip microcomputer contains an 8192-bit read-only memory programming, a 256-bit random-access memory for data storage, and a 4-bit binary arithmetic and logic unit. The chip, TMS1000, also has an oscillator activated by an external resistor and capacitor which, at a nominal frequency of 500kHz, provides an instruction cycle time of 12 μ s.

WW352 for further details Texas

Blue/u.v. photodetectors

A new series of photodiodes offer a spectral response between 250 and 550nm. The three devices in the range are planar diffused, oxide passivated silicon diodes with a response time of 0.5 μ s.

WW353 for further details Techmation

Variable shift register

The MC14557CP is a 1 to 64 bit c.m.o.s. variable length shift register that can be used for variable digital delay lines or to produce an odd length shift register. The device can operate at 8MHz from a 10V supply in the temperature range -40 to +85°C.

WW354 for further details GDS

Hall-effect switches

Two Hall-effect solid state switches designated ULN3006 and UL3006T are now available housed in transistor packages. The devices consist of a silicon Hall generator, amplifier, trigger and output stage integrated with a voltage regulator on a monolithic chip. The former device has an operating temperature range from 0 to +70°C

while the latter switch operates in the range from -40 to +150°C.

WW355 for further details Sprague

Variable frequency source

The MC14411 c.m.o.s. generator provides 16 clock frequencies which are available simultaneously. The device contains a crystal oscillator, a programmable rate-select circuit, and divider chains. When connected to a 1.8432MHz \pm 0.05% crystal, 14 frequencies from 75Hz to 9600Hz are produced. The oscillator frequency and a signal at half the crystal frequency provide the other two outputs.

WW356 for further details Motorola

Interface i.c.

AMI Microsystems have introduced the S1883 — a single chip m.o.s./l.s.i. universal asynchronous receiver/transmitter. The device is capable of transmitting and receiving in full duplex mode at data rates up to 12.5k baud.

WW357 for further details AMI

240V transistor

The type BUX80 power transistor is designed for use in high-frequency switched-mode power supplies operating from a 240V mains supply. A glass passivation construction is used in the transistor which is rated at 100W with a V_{CEO} of 400V.

WW358 for further details Mullard

16K r.o.m.

The RO-3-8316A is a high yield 8-bit word read-only memory which costs about £6 and is for use with microprocessors. The device operates from a single 5V supply and offers an access time of 850ns with a power dissipation of around 200mW.

WW359 for further details G.I.

Fast recovery rectifiers

A range of axial lead silicon power rectifiers from Semtech offer a 30ns reverse recovery time. The devices are claimed to be 98.5% efficient and are available with current ratings from 0.5 to 10A and p.i.v. ratings up to 150V.

WW360 for further details Bourns

275V f.e.t.s

A new family of junction f.e.t.s with interchangeable sources and drains have a guaranteed minimum breakdown voltage of 275V.

WW361 for further details Siliconix

Audio amplifier

An audio amplifier packaged in a SOT32 type case will deliver 10W into 4 Ω with a distortion figure of under 1% at 7W. The circuit incorporates frequency compensation and requires one external capacitor for complete stability.

WW362 for further details Texas

Count/display i.c.

Ferranti have recently added the 2N1040E to their range of i.c.s. This device is a universal count/display circuit that can be adapted to drive most types of display. The chip uses the collector diffusion isolation technique and can count from 0 to 5MHz in a forward or reverse direction.

WW363 for further details Edmundson

Telephone i.c.s

Plessey Semiconductors has introduced two i.c.s for telephone applications, the MP9100 push-button telephone dialler and the MP9200 repertory telephone store. The 9100 is a p-channel low threshold m.o.s. circuit containing the logic required to interface between a standard keyboard and a Strowger type telephone system.

The 9200 uses a similar construction and contains the logic and storage capability to form a self-contained repertory telephone number store of up to ten 22-digit numbers. The two devices can be used together to form a complete repertory dialling system.

WW364 for further details Plessey

SGS-ATES (UK) Ltd, Planar House, Walton Street, Aylesbury, Bucks, HP21 7QN.

Texas Instruments Ltd, Manton Lane, Bedford.

Techmation Ltd, 58 Edgware Way, Edgware, Middx HA8 8JP.

GDS Sales Ltd, Michaelmas House, Salt Hill, Bath Road, Slough, Bucks.

Sprague Electric (UK) Ltd, 159 High Street, Yiewsley, West Drayton, Middx.

Motorola Ltd, Semiconductor Products Division, York House, Empire Way, Wembley, Middx.

AMI Microsystems Ltd, 108A Commercial Road, Swindon, Wiltshire.

Mullard Ltd, Mullard House, Torrington Place, London WC1.

General Instrument Microelectronics Ltd, 57 Mortimer Street, London W1N 7TD.

Bourns (Trimpot) Ltd, Hodford House, 17 High Street, Hounslow, Middx TW3 1TE.

Siliconix Ltd, 30a High Street, Thatcham, Berks RG13 4JG.

Edmundson Electronic Components Ltd, 30 Ossory Road, London SE1 5AN.

Plessey Semiconductors, Cheney Manor, Swindon, Wilts SN2 2QW.



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Celestion present UL. An entirely new ultra modern range from the specialists who concentrate exclusively on the development and perfection of loudspeakers. Who continuously research new materials, new techniques and new processes in the constant endeavour to produce purer sounds, new subtleties and greater accuracy in the presentation of fine sound—for the benefit of the fastidious enthusiast.

The new UL range offers you a choice of three superb loudspeaker systems, each purpose-designed for its intended setting and appropriate ancillary equipment.

The **UL6**. Slim, sleek, discreet appearance, incorporating a new smoother tweeter unit extending beyond the limits of human hearing, a new mid-bass range unit employing a 1.5 inch (38mm) voice-coil, massive magnet

system and specially treated Bextrene diaphragm. Also, a new ABR (auxiliary bass radiator) which ensures excellent bass response, raises sensitivity and reduces distortion to very negligible limits.

The **UL8**. Next in order of size, cost and performance, has the new HD1000 tweeter, providing exceptional dispersion and complete freedom from 'listener fatigue' effects. The mid-bass has an 8 inch specially processed Bextrene diaphragm and 1.5 inch (38mm) voice-coil. The complementary ABR ensures superior response to bass notes. Results are equivalent to those normally obtained from substantially larger enclosures.

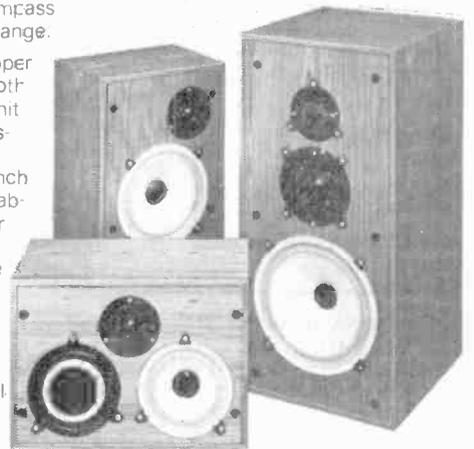
The **UL10**. A loudspeaker intended for use with associated amplifiers and signal sources of the highest quality. This truly excellent

loudspeaker will provide sound reproduction of the very highest standard, pleasing to the most discriminating listener. Three transducers are used to encompass the extreme frequency range.

The middle range and upper range transducers are both of soft-dome pressure-unit type—the low range transducer employs a treated Bextrene cone and 1.5 inch (38mm) voice-coil. An elaborate L/C crossover filter divides the frequency band to allow a response of exceptional smoothness.

The whole system is generously rated and when used under normal conditions, overloading is virtually impossible.

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Please send list of appointed Celestion UL dealers where I may hear a demonstration. I would also like a specification sheet.

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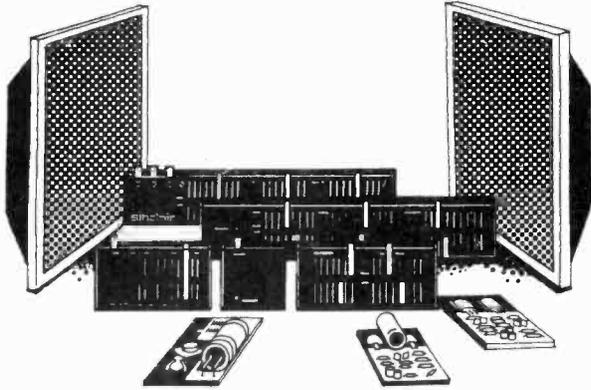
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Sinclair Project 80



The watts...

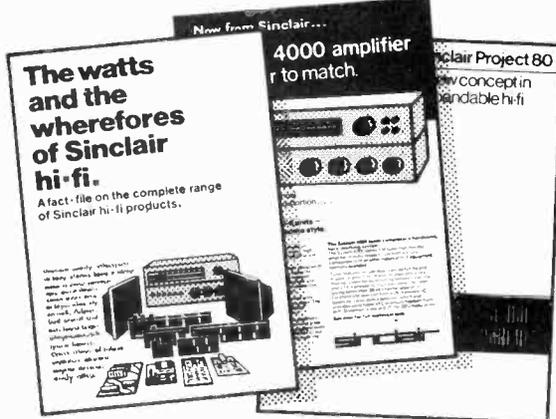
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To: Sinclair Radionics Ltd, FREEPOST, St Ives, Huntingdon, Cambs., PE17 4BR

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Marshall's

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TOP 200 IC'S TTL, CMOS & LINEARS

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CA3020A	1.80	CD4044	1.80	SL414	1.80	SN7450	0.16	SN74160	1.10
CA3028A	0.79	CD4045	2.85	SL610C	1.70	SN7451	0.16	SN74161	1.10
CA3035	1.37	CD4046	2.64	SL611C	1.70	SN7453	0.16	SN74162	1.10
CA3046	0.70	CD4047	1.85	SL612C	1.70	SN7454	0.16	SN74163	1.10
CA3048	2.11	CD4049	0.81	SL620C	2.80	SN7460	0.16	SN74164	2.01
CA3052	1.82	CD4050	0.86	SL621C	2.80	SN7470	0.33	SN74165	2.01
CA3089E	1.96	LM301A	0.48	SL623C	4.89	SN7472	0.26	SN74167	4.10
CA3090Q	4.23	LM308	2.80	SL640C	3.10	SN7473	0.36	SN74174	1.25
CD4000	0.36	LM05TL	1.80	SN7400	0.16	SN7474	0.36	SN74175	0.90
CD4001	0.36	LM380	1.10	SN7401	0.16	SN7475	0.80	SN74176	1.44
CD4002	0.36	LM381	2.20	SN7401AN	0.38	SN7476	0.35	SN74180	1.40
CD4006	1.58	LM702C	0.78	SN7402	0.16	SN7480	0.80	SN74181	1.96
CD4007	0.36	LM709	0.38	SN7403	0.16	SN7481	1.25	SN74190	2.30
CD4008	1.83	BD1L	0.45	SN7404	0.19	SN7482	0.75	SN74191	2.30
CD4009	1.18	14DIL	0.40	SN7405	0.19	SN7483	0.95	SN74192	1.15
CD4010	1.18	LM710	0.47	SN7406	0.45	SN7484	0.95	SN74193	1.15
CD4011	0.36	LM723C	0.90	SN7407	0.45	SN7485	1.25	SN74196	1.80
CD4012	0.36	LM741C	0.40	SN7408	0.19	SN7486	0.32	SN74197	1.58
CD4013	0.66	BD1L	0.40	SN7409	0.22	SN7490	0.45	SN74198	2.25
CD4014	1.72	14DIL	0.38	SN7410	0.16	SN7491	0.85	SN74199	2.25
CD4015	1.72	LM747	1.06	SN7411	0.25	SN7492	0.45	SN76003N	2.92
CD4016	0.66	LM748	0.60	SN7412	0.28	SN7493	0.45	SN76013N	1.95
CD4017	1.72	LM1401L	0.73	SN7413	0.35	SN7494	0.82	SN76023N	1.80
CD4018	2.56	LM3900	0.70	SN7416	0.35	SN7495	0.72	SN76033N	2.92
CD4019	0.86	MC1303L	1.60	SN7417	0.35	SN7496	0.75	TAA263	1.10
CD4020	1.91	LM7812	2.50	SN7420	0.16	SN74100	1.25	TAA300	2.80
CD4021	1.72	LM7815	2.80	SN7423	0.29	SN74107	0.36	TAA350A	1.10
CD4022	1.66	LM7824	2.50	SN7425	0.29	SN74118	1.00	TAA550	0.60
CD4023	0.66	MC1303L	1.60	SN7427	0.29	SN74119	1.92	TAA611C	2.18
CD4024	1.24	MC1310P	2.59	SN7430	0.16	SN74121	0.37	TAA621	2.03
CD4025	0.32	MC1330P	0.90	SN7432	0.28	SN74122	0.50	TAA661B	1.32
CD4027	0.43	MC1351P	0.80	SN7437	0.35	SN74123	0.60	TBA641B	2.25
CD4028	1.80	MC1352P	0.80	SN7438	0.35	SN74141	0.85	TBA651	1.69
CD4029	3.50	MC1466L	3.50	SN7440	0.18	SN74145	0.90	TBA800	1.40
CD4030	0.87	MC1469R	2.75	SN7444AN	0.85	SN74150	0.60	TBA810	1.40
CD4031	5.19	NE555V	0.70	SN7442	0.65	SN74151	0.85	TBA820	1.15
CD4037	1.93	NE556	1.30	SN7445	0.90	SN74153	0.85	TBA920	4.00
CD4041	1.86	NE560	4.48	SN7446	0.95	SN74154	1.50	DIL sockets	0.17
CD4042	1.38	NE561	4.48	SN7447	0.96	SN74155	1.50		

BRISTOL

10% discount for callers at our new shop at 1 Straits Parade, Fishponds, Bristol BS16 2LX during August. Tele: Bristol 654201/2.

POPULAR SEMICONDUCTORS

2N696	0.22	2N3906	0.27	AFJ39	0.85	BD139	0.71	MPSA56	0.31
2N697	0.16	2N4037	0.42	AF239	0.85	BD140	0.87	OC28	0.765
2N698	0.82	2N4036	0.67	AF240	0.90	BF115	0.36	OC35	0.80
2N699	0.59	2N4058	0.18	AF219	0.70	BF117	0.55	OC42	0.50
2N706	0.14	2N4062	0.15	AF280	0.79	BF154	0.20	OC45	0.32
2N708	0.17	2N4289	0.34	AL102	1.00	BF159	0.27	TIP29A	0.49
2N916	0.28	2N4920	1.10	BC107	0.14	BF180	0.35	TIP29C	0.58
2N918	0.32	2N4921	0.83	BC108	0.14	BF181	0.36	TIP31A	0.62
2N1302	0.185	2N4923	1.00	BC109	0.14	BF184	0.30	TIP32A	0.74
2N1304	0.26	2N5245	0.47	BC147B	0.14	BF194	0.12	TIP33A	1.01
2N1306	0.31	2N5294	0.48	BC148B	0.15	BF195	0.12	TIP34A	1.51
2N1308	0.47	2N5296	0.48	BC149B	0.15	BF196	0.13	TIP35A	2.90
2N1711	0.45	2N5457	0.49	BC157A	0.16	BF197	0.18	TIP36A	3.70
2N2102	0.80	2N5458	0.46	BC158A	0.16	BF198	0.18	TIP42A	0.90
2N2147	0.78	2N5459	0.49	BC167B	0.15	BF244	0.21	TIP295	0.98
2N2148	0.94	2N6027	0.45	BC168B	0.15	BF257	0.47	TIP305S	5.50
2N2218A	0.22	3N128	0.73	BC169B	0.15	BF258	0.53	TIS43	0.28
2N2219A	0.26	3N140	1.00	BC182	0.12	BF259	0.55	ZTX300	0.13
2N2220	0.25	3N141	0.81	BC182L	0.12	BF561	0.27	ZTX301	0.13
2N2221	0.18	3N200	2.49	BC183	0.12	BF598	0.25	ZTX500	0.16
2N2222	0.20	40361	0.40	BC183L	0.12	BF939	0.24	ZTX501	0.13
2N2369	0.20	40362	0.45	BC184	0.13	BF979	0.24	ZTX502	0.18
2N2646	0.55	40406	0.44	BC184L	0.13	BF929	0.30	1N914	0.07
2N2904	0.22	40407	0.36	BC212A	0.16	BF930	0.27	1N9754	0.16
2N2905	0.25	40408	0.30	BC212LA	0.16	BF981	0.24	1N4007	0.10
2N2906	0.19	40409	0.52	BC213LA	0.15	BF985	0.30	1N4148	0.07
2N2907	0.22	40410	0.52	BC214LB	0.16	BF988	0.25	1N5404	0.22
2N2924	0.20	40411	2.00	BC237B	0.18	BF950	0.225	1N5408	0.30
2N2926G	0.12	40594	0.74	BC238C	0.15	BF951	0.23	AA119	0.06
2N3053	0.25	40595	0.84	BC239C	0.15	BF952	0.205	BA102	0.25
2N3054	0.60	40636	1.10	BC257A	0.16	BR939	0.48	BA145	0.18
2N3055	0.75	40673	0.79	BC258B	0.16	ME0402	0.20	BA154	0.12
2N3391	0.28	AC126	0.20	BC259B	0.17	ME0412	0.18	BA155	0.12
2N3392	0.15	AC127	0.20	BC301	0.34	ME4102	0.11	BB1038	0.23
2N3393	0.16	AC128	0.20	BC307B	0.17	MJ480	0.96	BB1048	0.46
2N3440	0.59	AC151	0.27	BC308A	0.15	MJ481	1.20	BY126	0.12
2N3442	1.40	AC152	0.43	BC309C	0.20	MJ490	1.05	BY127	0.15
2N3638	0.15	AC153	0.35	BC327	0.23	MJ491	1.45	BY211	0.51
2N3702	0.12	AC176	0.30	BC328	0.22	MJ2955	1.00	BY212	0.51
2N3703	0.13	AC187K	0.35	BCY70	0.17	MJE340	0.48	OA47	0.06
2N3704	0.15	AC188K	0.40	BCY71	0.22	MJE370	0.65	OAB1	0.18
2N3706	0.15	AD143	0.68	BCY72	0.15	MJE371	0.75	OA90	0.06
2N3708	0.14	AD161	0.50	BD121	1.00	MJE520	0.60	OA91	0.06
2N3714	1.38	AD162	0.50	BD123	0.82	MJE521	0.70	WO21A200	0.32
2N3716	1.80	AF106	0.40	BD124	0.87	MJE2955	1.20	BY164	0.57
2N3771	2.20	AF109	0.40	BD131	0.40	MJE3055	0.75	STZ disc	0.20
2N3773	2.65	AF115	0.35	BD132	0.35	MJE8112	0.47	40699	1.00
2N3789	2.06	AF116	0.35	BD135	0.43	MPE102	0.39	TIC44	0.29
2N3819	0.37	AF117	0.35	BD136	0.47	MPSA05	0.25	C106D	0.65
2N3820	0.64	AF118	0.35	BD137	0.55	MPSA06	0.31	ORP2	0.60
2N3904	0.27	AF124	0.30	BD138	0.63	MPSA55	0.31		

Prices correct at July, 1975, but all exclusive of V A T

Post & Package 25p

S-2020TA STEREO TUNER/AMPLIFIER KIT

NEW PRODUCT

A high-quality push-button FM Varicap Stereo Tuner combined with a 20W r.m.s. per channel Stereo Amplifier.

Brief Spec. Amplifier: Low field Toroidal transformer, Mag. input, Tape In/Out facility (for noise reduction unit, etc). THD less than 0.1% at 20W into 8 ohms. All sockets, fuses, etc. are PC mounted for ease of assembly. Tuner section: uses Mullard LP1186 module requiring no RF alignment, ceramic IF, INTERSTATION MUTE, and phase-locked IC stereo decoder. LED tuning and stereo indicators. Tuning range 88–104MHz. 30dB mono S/N @ 1.8µV. THD typ. 0.4%.

PRICE: £47.95 + 99p p&p + VAT.



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NELSON-JONES STEREO FM TUNER

A very high performance tuner with dual gate MOSFET RF and Mixer front end, triple gang varicap tuning, and dual ceramic filter/dual IC IF amp.

Brief Spec. Tuning range 88–104MHz. 20dB mono quieting @ 0.75µV. Image rejection—70dB. IF rejection—85dB. THD typically 0.4%.

IC stabilized PSU and LED tuning indicators. Push-button tuning and AFC unit. Choice of either mono or stereo with a choice of stereo decoders.

PRICE: Mono £25.46 + 85p p&p + VAT;

With Portus-Haywood Decoder £31.96 + 85p p&p + VAT;

With ICPL Decoder £29.73 + 85p p&p + VAT.

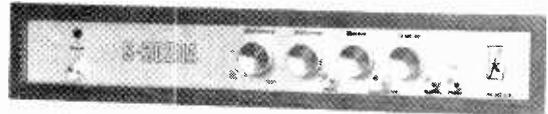
NEW PRODUCT

S-2020A AMPLIFIER KIT

Developed in our laboratories from the highly successful "TEXAN" design. PC mounting potentiometers, switches, sockets and fuses are used for ease of assembly and to minimize wiring.

Typ. Spec. 20+20W r.m.s. into 8-ohm load at less than 0.1% THD. Mag. PU input S/N 60dB. Radio input S/N 72dB. Head-phone output. Tape In/Out facility (for noise reduction unit, etc). Toroidal mains transformer.

PRICE: £29.95 + 99p p&p + VAT.



STEREO MODULE TUNER

A low-cost Stereo Tuner based on the Mullard LP1186 RF module requiring no alignment. The IF comprises a ceramic filter and high-performance IC. Variable INTERSTATION MUTE. PLL stereo decoder IC.

Typ. Spec. Sens. 30dB S/N mono @ 1.8µV. Tuning range 88–104MHz. LED sig. strength indicator. LED Stereo indicator. THD typically 0.4%.

PRICE: Stereo £26.32 + 85p p&p + VAT. Mono £22.40 + 85p p&p + VAT.

ALL THE ABOVE KITS ARE SUPPLIED COMPLETE WITH ALL METALWORK, SOCKETS, FUSES, NUTS AND BOLTS, KNOBS, FRONT PANELS, SOLID MAHOGANY CABINETS AND COMPREHENSIVE INSTRUCTIONS.

SUB ASSEMBLIES

BASIC NELSON-JONES TUNER

Supplied as a printed circuit board with all components and screening box to build a varicap tuner module. Performance spec as above for complete N-J Tuner. For suitable stereo decoders see below. (Illustrated without screening box.)

PRICE: £12.88 + 25p p&p + VAT.



BASIC MODULE TUNER

Supplied as a printed circuit board with all components and screened Mullard LP1186, to build a mono or stereo tuner module. Performance spec as above for Stereo Module Tuner complete kit.

PRICE: Mono £11.11 + 25p p&p + VAT; Stereo £13.89 + 25p p&p + VAT.



PORTUS-HAYWOOD PHASE-LOCKED STEREO DECODER

Mk II version of this design (WW Sept. 1970). The lowest distortion phase-locked stereo decoder kit available (Typ. 0.05% @ N-J Tuner O/P level). Separation 40dB up to 15KHz.

Complete kit comprises PCB and all components, inc. stereo LED.

PRICE: £7.68 + 25p p&p + VAT.



PHASE-LOCKED IC DECODER

Integrated circuit phase-locked stereo decoder based on the MC1310. THD typically 0.3%. Separation 40dB @ 1KHz.

PRICE: £4.27 + 20p p&p + VAT.



PUSH-BUTTON UNIT

The six-position push-button unit used in our tuners and tuner/amp. Each track has the required diode law for stability of tuning. There are approx. 40 turns on each button and there are six separate moving pointers. An AFC disable switch is incorporated with each button. The unit is finished in black with red pointers.

PRICE: £3.00 + 20p p&p + VAT.



Please send SAE for complete lists and specifications.

INTEGREX LIMITED, Portwood Industrial Estate, Church Gresley, Burton-on-Trent, Staffs, DE11 9PT.
Tel. Swadlincote (0283 87) 5432. Telex 377106.

TRANSFORMERS

CASED TRANSFORMERS

Housed in smart resin coated steel cases with 3 core power cable and outlet socket. fused primary winding. Isolation types are fitted with 3-pin outlet sockets and are available with 110 volt or 240 volt output (Please state). Auto types are fitted with 2-pin flat style sockets up to 500 VA. 3-pin sockets from 750 to 3000 VA. See Auto and Isolation sections for prices. Plugs Extra.



SAFETY ISOLATING

Prim. 120/240V. Sec. 120/240V. Centre Tap with screen

VA (WATTS)	REF No.	PRICE Cased £	PRICE 2 Pin + 1 Earth £	PRICE Open £	Post £
60	149	8.35	0.88	4.37	0.56
100	150	9.15	0.88	4.90	0.64
200	151	11.45	0.88	6.14	0.80
250	152	12.90	0.88	6.80	0.86
350	153	15.50	0.88	11.88	0.95
500	154	17.25	0.88	13.62	1.13
750	155	27.10	1.10	20.59	0.4.
1000	156	35.40	1.10	29.15	0.4.
1500	157	42.00	1.10	33.37	0.4.
2000	158	49.75	2.64	37.10	0.4.
3000	159	73.15	2.64	58.55	0.4.

MINIATURE & EQUIPMENT

Primary 240V with Screen

VOLTS		MILLIAMPS		PRICE £	Post £
Sec. 1	Sec. 2	Sec. 1	Sec. 2		
3.0-3	200	200	238	1.50	0.25
0.6	0.6	500	500	234	1.38
0.6	0.6	1000	1000	212	1.90
9.0-9	—	100	—	13	1.40
0.9	0.9	330	330	235	1.50
0.9-9	0.9-9	500	500	207	1.93
0.9-9	0.9-9	1000	1000	208	2.75
15.0-15	—	40	—	240	1.35
0.15	0.15	200	200	236	1.38
20.0-20	—	30	—	241	1.35
0.20	0.20	150	150	237	1.38
0.15-20	0.15-20	500	500	205	2.73
0.20	0.20	300	300	214	1.93
0.20	—	3500	No Screen	1116	3.30
20.12-0	—	700	—	221	2.20
12.20	(D.C.)	—	—	—	—
0.15-20	0.15-20	1000	1000	206	3.50
0.15-27	0.15-27	500	500	203	3.00
0.15-27	0.15-27	1000	1000	204	3.85

12 and 24 VOLTS PRIMARY 200-240 Volts

VA (Watts)	TYPE	PRICE £	Post £
12V	24V		
0.3	0.15	242	1.58
0.5	0.25	111	1.38
1	0.5	213	1.74
2	1	71	2.30
4	2	18	2.96
6	3	70	4.18
8	4	106	4.56
10	5	72	5.20
12	6	116	5.51
16	8	17	7.00
20	10	115	10.42
30	15	187	13.25
40	20	232	14.85
60	30	226	16.83

30 VOLTS

PRIMARY 200/240V

SECONDARY 12, 15, 20, 24, 30V

AMPS	Ref.	Price £	Post £
0.5	112	1.90	0.47
1	79	2.40	0.56
2	3	3.50	0.56
3	20	4.50	0.64
4	21	5.15	0.72
5	51	6.40	0.72
6	117	7.16	0.88
8	88	9.55	0.95
10	89	9.67	0.95

50 VOLTS

PRIMARY 200/240V

SECONDARY 19, 25, 33, 40, 50V

AMPS	Ref.	Price £	Post £
0.5	102	2.58	0.47
1	103	3.48	0.56
2	104	5.03	0.64
3	105	5.81	0.72
4	106	7.58	0.88
6	107	12.30	0.95
8	118	13.20	1.13
10	119	17.02	0.4.

60 VOLTS

PRIMARY 200/240V

SECONDARY 24, 30, 48, 60V

AMPS	Ref.	Price £	Post £
0.5	124	2.30	0.56
1	126	3.41	0.56
2	127	5.09	0.72
3	125	7.52	0.80
4	123	8.75	0.95
5	40	9.75	0.95
6	120	11.20	1.01
8	121	15.00	1.19
10	122	18.20	0.4.
12	189	18.50	0.4.

AUTO TRANSFORMERS

VA (Watts)	Ref No	PRICE Cased £	PRICE Plugs 2 & 3 pin £	PRICE Open £	Post £
Tapped at 115, 220, 240 Volts					
20	113	3.85	0.20	1.71	0.47
Tapped at 115, 200, 220, 240 Volts					
150	4	6.38	0.20	4.12	0.56
200	65	7.04	0.20	4.95	0.64
300	66	8.00	0.20	5.81	0.72
500	67	10.99	0.20	8.85	0.88
750	83	13.82	0.85	10.80	0.95
1000	84	17.27	0.85	13.68	1.13
1500	93	21.87	0.85	18.31	0.4.
2000	95	33.11	1.60	24.25	0.4.
3000	73	47.94	2.10	35.10	0.4.

BRIDGE RECTIFIERS

ONE AMP	Price
50 P.I.V.	0.20
100 P.I.V.	0.20
200 P.I.V.	0.28
600 P.I.V.	0.30



FOUR AMP	Price
100 P.I.V.	0.55
200 P.I.V.	0.59
400 P.I.V.	0.65
600 P.I.V.	0.75

TWO AMP	Price
100 P.I.V.	0.35
100 P.I.V.	0.40
200 P.I.V.	0.45
400 P.I.V.	0.50

SIX AMP	Price
50 P.I.V.	0.65
100 P.I.V.	0.70
200 P.I.V.	0.08
400 P.I.V.	0.90

POWER UNIT TYPE CC12-05



Output switched 3, 4.5, 6, 7.5, 9 and 12 Volts at 500 mA D.C. Operates from 240 V mains, suitable for Radios, Tape Recorders, Record Players, etc. Size 7.5 x 5.0 x 14.0 cm. Price £3.95 Post 25p.

NEW! 2" AND 4" PANEL METERS

2"		4"	
SIZE: 60mm Wide x 45mm High x 40mm Deep.	Movement	SIZE: 110mm Wide x 82mm High x 43mm Deep.	Movement
0-50 micro A	1250	0-50 micro A	1400
0-100 micro A	580	0-100 micro A	730
0-500 micro A	170	0-500 micro A	220
0-1 mA	170	0-1 mA	200
0-5 mA	170	0-5 mA	200
0-10 mA	6	0-10 mA	6
0-50 mA	0.5	0-50 mA	0.5
0-100 mA	0.5	0-100 mA	0.5
0-500 mA	0.5	0-500 mA	0.5
0-1 AMP	0.5	0-1 AMP	0.5
0-2 AMP	0.5	0-2 AMP	0.5
0-25 Volt	15K	0-25 Volt	15K
0-50 Volt	50K	0-50 Volt	50K
0-300 Volt	300K	0-300 Volt	300K
1/2 Meter	9250	1/2 Meter	200
VU Meter	9250	VU Meter	5250

VU Meters are complete with detectors. Modern wide view. Price 2" £3.20 Post 10p. Price 4" £4.00 Post 10p. Lamps 60p per set.

C1000 MULTI-METER

Compact General Purpose Mini Multimeter
Input Resistance 1000 ohms per volt
Ranges: AC Volts 0.15 50 250 1000 Volts
DC Volts 0.10 50 250 1000 Volts
DC Current 0-1 mA 0-100 mA
Resistance 0-150K ohms
Size 60 x 24 x 90 mm
Complete with Batteries, Test Prods. Instructions



Special price £3.30 Post 25p

1/4-WATT CARBON FILM RESISTORS

also available 1/4 watt at 70°C 12 range 100—1MΩ. 5% tol above 470KΩ. 10% tol at 96p per 100

MINIATURE NEONS

6mm dia. 12mm length leads length approx. 20mm. Recommended ballast resistor 150K ohms for 240 Volt operation. Price: Packed 10 for 80p. Postage 15p.

PLEASE ADD VAT

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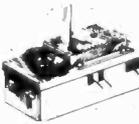
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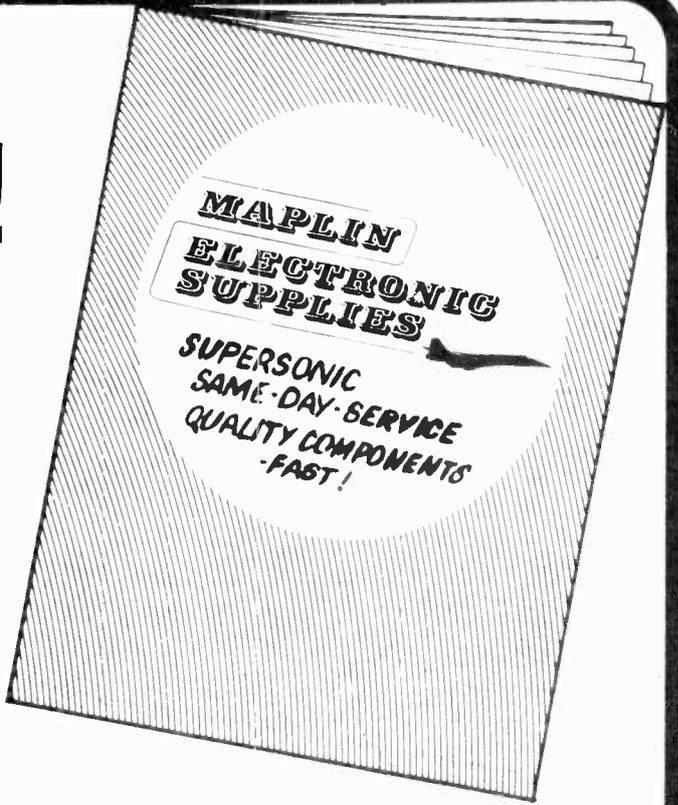
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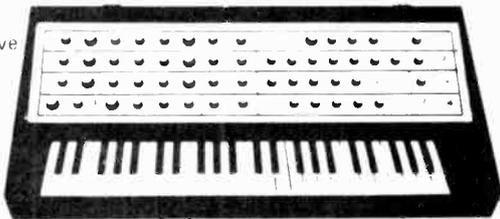
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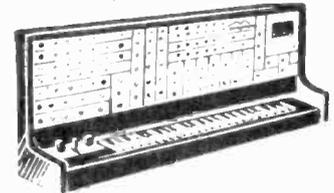
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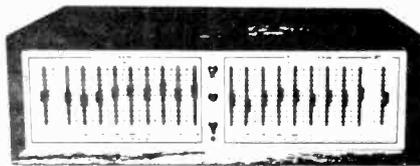
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ECC83	0.38	EM84	0.40	HL4	0.50	UCL83	1.50	6K8GT	0.50	30PL33	1.10
ECC85	0.45	EM85	0.45	HL4	0.50	UCL83	1.50	6K8GT	0.50	30PL34	1.10
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ECC90	0.45	EM87	0.45	HL4	0.50	UCL83	1.50	6K8GT	0.50	30PL36	1.10
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1B35A	5B/257M	715B	CV135	CV3998	EF80A	Q24	Q24A	Q2400	Q2403
1B36A	5C/22	723A/B	CV136	CV4001	EL91	Q2400	Q2403	Q2404	Q2406
1N21	5C/22	725A	CV137	CV4002	EN30	Q2406	Q2407	Q2408	Q2409
1N21B	5D21	725A	CV138	CV4003	EN31	Q2410	Q2411	Q2412	Q2413
1N21B	5D48Y	725A	CV140	CV4004	EN32	Q2414	Q2415	Q2416	Q2417
1N23B	5U48Y	725A	CV144	CV4005	EN91	Q2418	Q2419	Q2420	Q2421
1N23CR	5Z3	725A	CV160	CV4006	ESU74	Q2422	Q2423	Q2424	Q2425
1X2B	5Z4G	725A	CV173	CV4007	ESU76	Q2426	Q2427	Q2428	Q2429
2A3	6A/4A	807	CV187	CV4008	ESU77	Q2430	Q2431	Q2432	Q2433
2A515	6A/5	808	CV188	CV4009	F6057	Q2434	Q2435	Q2436	Q2437
2A515	6A/5	811	CV190	CV4010	F6060	Q2438	Q2439	Q2440	Q2441
2A515	6A/5	811A	CV191	CV4011	F6061	Q2442	Q2443	Q2444	Q2445
2A515	6A/5	812A	CV220	CV4012	F6062	Q2446	Q2447	Q2448	Q2449
2A515	6A/5	813	CV219	CV4013	F6063	Q2450	Q2451	Q2452	Q2453
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2A515	6A/5	818	CV286	CV4015	F6065	Q2458	Q2459	Q2460	Q2461
2A515	6A/5	828	CV287	CV4016	F6066	Q2462	Q2463	Q2464	Q2465
2A515	6A/5	829B	CV288	CV4017	F6067	Q2466	Q2467	Q2468	Q2469
2A515	6A/5	830B	CV289	CV4018	F6068	Q2470	Q2471	Q2472	Q2473
2A515	6A/5	830B	CV290	CV4019	F6069	Q2474	Q2475	Q2476	Q2477
2A515	6A/5	830B	CV332	CV4020	F6070	Q2478	Q2479	Q2480	Q2481
2A515	6A/5	830B	CV342	CV4021	F6071	Q2482	Q2483	Q2484	Q2485
2A515	6A/5	830B	CV343	CV4022	F6072	Q2486	Q2487	Q2488	Q2489
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2A515	6A/5	830B	CV349	CV4028	F6078	Q2510	Q2511	Q2512	Q2513
2A515	6A/5	830B	CV350	CV4029	F6079	Q2514	Q2515	Q2516	Q2517
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2A515	6A/5	830B	CV352	CV4031	F6081	Q2522	Q2523	Q2524	Q2525
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2A515	6A/5	830B	CV355	CV4034	F6084	Q2534	Q2535	Q2536	Q2537
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2A515	6A/5	830B	CV365	CV4044	F6094	Q2574	Q2575	Q2576	Q2577
2A515	6A/5	830B	CV366	CV4045	F6095	Q2578	Q2579	Q2580	Q2581
2A515	6A/5	830B	CV367	CV4046	F6096	Q2582	Q2583	Q2584	Q2585
2A515	6A/5	830B	CV368	CV4047	F6097	Q2586	Q2587	Q2588	Q2589
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2A515	6A/5	830B	CV370	CV4049	F6099	Q2594	Q2595	Q2596	Q2597
2A515	6A/5	830B	CV371	CV4050	F6100	Q2598	Q2599	Q2600	Q2601
2A515	6A/5	830B	CV372	CV4051	F6101	Q2602	Q2603	Q2604	Q2605
2A515	6A/5	830B	CV373	CV4052	F6102	Q2606	Q2607	Q2608	Q2609
2A515	6A/5	830B	CV374	CV4053	F6103	Q2610	Q2611	Q2612	Q2613
2A515	6A/5	830B	CV375	CV4054	F6104	Q2614	Q2615	Q2616	Q2617
2A515	6A/5	830B	CV376	CV4055	F6105	Q2618	Q2619	Q2620	Q2621
2A515	6A/5	830B	CV377	CV4056	F6106	Q2622	Q2623	Q2624	Q2625
2A515	6A/5	830B	CV378	CV4057	F6107	Q2626	Q2627	Q2628	Q2629
2A515	6A/5	830B	CV379	CV4058	F6108	Q2630	Q2631	Q2632	Q2633
2A515	6A/5	830B	CV380	CV4059	F6109	Q2634	Q2635	Q2636	Q2637
2A515	6A/5	830B	CV381	CV4060	F6110	Q2638	Q2639	Q2640	Q2641
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2A515	6A/5	830B	CV383	CV4062	F6112	Q2646	Q2647	Q2648	Q2649
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CD4099AE	£2.95	£2.46	£1.96

	1-24	25-99	100+
7400	14p	12p	10p
7401	14p	12p	10p
7402	14p	12p	10p
7403	15p	12p	10p
7404	16p	13p	11p
7408	16p	13p	11p
7409	16p	13p	11p
7410	16p	13p	11p
7413	29p	24p	20p
7417	27p	22p	20p
7420	16p	13p	11p
7427	27p	22p	18p
7430	16p	13p	11p
7432	27p	22p	18p
7437	27p	22p	18p
7441	75p	62p	50p
7442	65p	55p	45p
7445	85p	71p	57p
7447	95p	83p	67p
7447A	95p	83p	67p
7448	85p	71p	57p
7470	30p	25p	20p
7472	25p	21p	17p
7473	30p	25p	20p
7474	32p	26p	21p
7475	47p	39p	31p
7476	32p	26p	21p
7482	75p	62p	50p
7485	£1.30	£1.09	87p
7486	32p	26p	21p
7489	£3.56	£2.80	£2.10
7490	49p	40p	32p
7491	65p	55p	45p
7492	57p	46p	36p
7493	49p	40p	32p
7495	67p	55p	45p
74100	£1.08	89p	72p
74107	35p	28p	22p
74121	34p	28p	23p
74122	47p	39p	31p
74141	78p	63p	53p
74145	68p	58p	48p
74154	£1.75	£1.48	80p
74174	£1.00	83p	67p
74180	£1.06	88p	71p
74181	£3.20	£2.50	£1.90
74192	£1.35	£1.14	90p
74193	£1.35	£1.14	90p
74196	£1.64	£1.34	99p

555 (8 pin dip) V	55p
555 (TO-99) T	81p
556 (14 pin dip)	£1.29
703 (RF/IF Amp)	68p
709 (8 pin dip)	38p
709 (TO-99)	45p
709 (14 pin dip)	39p
710 (8 pin dip)	39p
710 (TO-99)	45p
710 (14 pin dip)	44p
711 (TO-99)	51p
711 (14 pin dip)	44p
720 (A.M. Radio)	£1.76
723 (TO-99)	£1.09
723 (14 pin dip)	74p
741 (8 pin dip)	36p
741 (TO-99)	43p
741 (14 pin dip)	36p
747 (14 pin dip)	£1.04
748 (8 pin dip)	42p
748 (TO-99)	46p
748 (14 pin dip)	49p
753 (F.M. 1st. I.F.)	£1.08
75491	88p
75492	£1.10
Regulators 100 mA	
78105WC (TO-92)	60p
7812WC (TO-92)	60p
7815WC (TO-92)	60p
Regulators 100mA	
78105AWC (TBA625A)90p	
7812AWC (TBA625B)90p	
7815AWC (TBA625C)90p	
Regulators 500mA	
78M05HC	£1.35
78M12HC	£1.35
78M15HC	£1.35
78M18HC	£1.35
78M24HC	£1.35
Regulators 1A	
7805K (TO-3)	£2.09
7812K (TO-3)	£2.09
7815K (TO-3)	£2.09
7818K (TO-3)	£2.09
7824K (TO-3)	£2.09
Regulators 1A	
7805UC (TO-220)	£1.72
7812UC (TO-220)	£1.72
7815UC (TO-220)	£1.72
7818UC (TO-220)	£1.72
7824UC (TO-220)	£1.72
ICL8038	£3.52
AY-1-0212	£6.93
AY-1-5051	£1.44
AY-5-1224	£3.95
AY-5-3500	£6.59
AY-5-3507	£6.59
AY-5-4007	£7.94

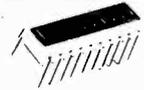
BHA0002	£3.01
CA2111	£1.19
CA3045	£1.69
CA3046	88p
CA3053	59p
CA3065	£1.60
CA3075	£1.64
CA3078	£1.26
CA3080	59p
CA3081	£1.86
CA3082	£1.86
CA3089E (TDA1200)	£2.43
CA3097E	£1.67
CA3123E	£1.76
CA3401E (LM390C)	68p
CA3600E	£1.44
MFC6040	87p
MFC6060A	79p
MFC6030A	79p
MFC6040	96p
MFC6070	£1.66
CT7001	£5.34
LM005T1 (TO-3)	£1.46
LM036T1 (TO-3)	£1.46
LM037T1 (TO-3)	£1.46
LM129 (SOT-32)	85p
LM130 (SOT-32)	85p
LM131 (SOT-32)	85p
LM301 T (TO-99)	65p
LM301 S (E pin dip)	59p
LM301A 1 (TO-99)	67p
LM301A S (8 pin dip)	59p
LM301 T (TO-99)	59p
LM307 S (8 pin dip)	57p
LM308 T (TO-99)	£1.96
LM308 S (8 pin dip)	98p
LM308A 1 (TO-99)	£7.92
LM308A S (8 pin dip)	£6.90
LM309K	£2.34
LM339	£2.25
LM370N	£2.85
LM371	£2.08
LM372N	£1.99
LM373N	£2.99
LM377N	£2.71
LM380	£1.25
LM381	£1.85
LM382	£1.66
LM703	68p
LM1820	£1.03
LM2111	£1.12
LM3900	69p
MC1303L	£1.84
MC1306P	80p
MC1310P	£2.39
MC1312	£2.42
MC1314	£4.13
MC1315	£4.62
MC1327	£1.12p
MC1330P	83p
MC1339P	£1.52
MC1350	64p
MC1351	88p
MC1352	88p
MC1357	£1.52
MC1358 (CA3065)	£1.16
MC1375	£1.48
MC1455 (555T)	62p
MC1456CG	£1.68
MC1458CPI	84p
MC1468C	£2.18
MC1495L	£4.24
MC1496G	96p
MC3302P	£1.50
MC3461P	74p
MMS314	£4.80
MMS316	£9.99
MVR5V (TO-3)	£1.45
MVR12V (TO-3)	£1.45
MVR15V (TO-3)	£1.45
NE540L	£1.25
NE546A	£1.16
NE555V	73p
NE556	£1.29
NE560B	£5.06
NE561B	£5.06
NE562B	£5.06
NE563	£2.96
NE565N	£2.63
NE566V	£1.87
NE567V	£2.63
SL414A	£2.09
SL415A	£2.75
SL437D	£7.50
SL44C	£2.84
SL610C	£2.03
SL611C	£2.03
SL612C	£2.03
SL613C	£4.31
SL620C	£3.06
SL621C	£3.06
SL622C	£7.62
SL623C	£5.57
SL624C	£2.84
SL630C	£1.87
SL640C	£3.75
SL641C	£3.75
SL645C	£3.75
SL650C	£9.85
SN745491N	88p
SN745492N	£1.10
SN76001N (TAA611)	£1.82
SN76003N	£3.30
SN76013N	£1.99
SN76023N	£1.98
SN76227N (MC1327)	£1.89
SN76532N	£1.86
SN76544N	£1.81
SN76550-2 (TAA550)	89p
SN76552-2	81p
SN76660N (TBA120)	75p
SN76666N (CA3065)	£1.12
TAA263	£1.50
TAA300	£2.16
TAA310A	£1.87
TAA320	£1.44
TAA350	£2.43
TAA370	£3.45
TAA550	75p
TAA570	£2.74
TAA700	£5.03
TBA1205	£1.25
TBA231	£1.02
TBA281 (723)	£2.59
TBA500Q	£3.16
TBA520Q	£3.85
TBA530Q	£3.27
TBA540Q	£3.72
TBA550Q	£5.29
TBA560CQ	£5.29
TBA625A	£1.03
TBA625B	£1.03
TBA625C	£1.03
TBA651	£1.87
TBA720Q	£2.79
TBA750Q	£2.79
TBA800	£1.11
TBA8105	£1.24
TBA810A5	£1.24
TBA820	80p
TBA920Q	£4.71
TBA990Q	£4.71
TCA270Q	£5.24
TCA760	£2.16
TCA800Q	£7.24
TCA8305	£1.04
TCA940	£2.25
TDA1054	£1.50
TDA1200	£2.43
TDA1405	80p
TDA1412	80p
TDA1415	80p
TDA2010	£3.00
TDA2020	£3.75
ULN2111A	£1.52
ZN414	£1.26

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RED	DL702R	DL707	DL701	DL704	£1.82
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RED	MAN71	MAN72	MAN73	MAN74	£1.82
YELLOW	MAN81	MAN82	MAN83	MAN84	£1.82
ORANGE	MAN3610	MAN3620	MAN3630	MAN3640	£1.82
GREEN	XAN51	XAN52	-	XAN54	£1.49
RED	XAN71	XAN72	-	XAN74	£1.49
YELLOW	XAN81	XAN82	-	XAN84	

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7400 Series TTL

	1	25	100+
SN7400	0.14	0.13	0.12
SN7401	0.14	0.13	0.12
SN7402	0.14	0.13	0.12
SN7403	0.14	0.13	0.12
SN7404	0.15	0.14	0.13
SN7405	0.15	0.14	0.13
SN7406	0.30	0.29	0.28
SN7407	0.30	0.29	0.28
SN7408	0.15	0.13	0.12

SN7409	0.15	0.13	0.12
SN7410	0.14	0.13	0.12
SN7411	0.23	0.22	0.21
SN7412	0.19	0.18	0.17
SN7413	0.30	0.29	0.28
SN7414	0.71	0.70	0.69
SN7415	0.30	0.29	0.27
SN7416	0.28	0.27	0.26
SN7417	0.28	0.27	0.26
SN7420	0.14	0.13	0.12
SN7421	0.95	0.94	0.93
SN7422	0.25	0.24	0.23
SN7423	0.26	0.25	0.22
SN7425	0.26	0.25	0.22
SN7426	0.26	0.25	0.22
SN7427	0.26	0.25	0.22
SN7428	0.39	0.38	0.37
SN7430	0.14	0.13	0.12
SN7432	0.25	0.24	0.22
SN7433	0.36	0.35	0.34
SN7437	0.27	0.26	0.22
SN7438	0.27	0.26	0.22
SN7439	1.10	1.08	1.06
SN7440	0.14	0.13	0.12
SN7441	0.70	0.69	0.66
SN7442	0.63	0.60	0.53
SN7443	1.00	0.99	0.90
SN7444	1.08	1.07	1.05

C-MOS Types

	1	25	100+
4000AE	0.24	0.19	0.16
4001AE	0.24	0.19	0.16
4002AE	0.24	0.19	0.16
4004AE	4.90	4.80	4.70
4006AE	1.55	1.50	1.45
4007AE	0.24	0.19	0.16
4008AE	1.60	1.55	1.50
4009AE	0.35	0.34	0.30
4010AE	0.50	0.48	0.46
4011AE	0.24	0.19	0.16
4012AE	0.24	0.19	0.16
4013AE	0.59	0.47	0.39
4014AE	1.65	1.33	1.10
4015AE	1.65	1.33	1.10
4016AE	0.62	0.50	0.41
4017AE	1.62	1.31	1.08
4018AE	2.45	2.40	2.30
4019AE	0.62	0.50	0.41
4020AE	1.82	1.46	1.21
4021AE	1.65	1.33	1.10
4022AE	1.65	1.33	1.10
4023AE	0.24	0.19	0.16
4024AE	1.13	0.91	0.75
4025AE	0.24	0.19	0.16
4026AE	6.50	6.40	6.30
4027AE	0.42	0.41	0.40
4028AE	1.41	1.14	0.94
4029AE	1.95	1.57	1.30
4030AE	0.62	0.50	0.41
4033AE	2.80	2.70	2.60
4035AE	1.71	1.38	1.14
4040AE	1.82	1.46	1.21
4041AE	0.90	0.85	0.80
4042AE	1.00	0.90	0.80
4043AE	1.00	0.90	0.80
4044AE	1.00	0.90	0.80
4047AE	1.50	1.45	1.40
4048AE	1.30	1.25	1.20
4049AE	0.54	0.44	0.36
4050AE	0.54	0.44	0.36
4051AE	1.77	1.43	1.18
4052AE	2.00	1.80	1.80
4053AE	2.70	2.60	2.50
4055AE	2.60	2.50	2.40
4056AE	1.35	1.09	0.90
4060AE	2.25	1.82	1.50
4066AE	0.95	0.90	0.85
4069AE	0.38	0.30	0.25
4071AE	0.34	0.27	0.23
4076AE	1.45	1.40	1.35

4081AE	0.34	0.27	0.23
4510AE	1.75	1.70	1.65
4516AE	1.76	1.42	1.18
4518AE	2.17	1.75	1.45
4520AE	1.65	1.60	1.50
4901AE	0.37	0.35	0.33
4911AE	0.37	0.35	0.33

LOW-POWER SCHOTTKY TTL

	1	25	100+
SN74LS00	0.58	0.56	0.54
SN74LS01	0.58	0.56	0.54
SN74LS02	0.58	0.56	0.54
SN74LS03	0.58	0.56	0.54
SN74LS04	0.63	0.61	0.59
SN74LS05	0.63	0.61	0.59
SN74LS08	0.58	0.56	0.54
SN74LS09	0.58	0.56	0.54
SN74LS10	0.58	0.56	0.54
SN74LS11	0.58	0.56	0.54
SN74LS15	0.58	0.56	0.54

SN74LS20	0.58	0.56	0.54
SN74LS21	0.58	0.56	0.54
SN74LS22	0.58	0.56	0.54
SN74LS27	0.64	0.62	0.60
SN74LS30	0.58	0.56	0.54
SN74LS32	0.64	0.62	0.60
SN74LS51	0.58	0.56	0.54
SN74LS54	0.58	0.56	0.54
SN74LS55	0.58	0.56	0.54
SN74LS73	0.92	0.90	0.88
SN74LS74	0.92	0.90	0.88

SN74LS76	0.92	0.90	0.88
SN74LS78	0.92	0.90	0.88
SN74LS107	0.92	0.90	0.88
SN74LS109	0.92	0.90	0.88
SN74LS112	0.92	0.90	0.88
SN74LS113	0.92	0.90	0.88
SN74LS114	0.92	0.90	0.88
SN74LS138	2.38	2.32	2.26
SN74LS139	2.38	2.32	2.26
SN74LS151	2.10	2.05	2.00
SN74LS153	2.38	2.32	2.26

SN74LS1	0.34	0.33	0.30
SN74LS3	0.74	0.71	0.68
SN74LS4	0.89	0.87	0.80
SN74LS9	1.62	1.58	1.50
SN74LS9	1.74	1.71	1.65
SN74LS9	1.62	1.58	1.50
SN74LS9	1.50	1.45	1.40
SN74LS9	1.60	1.55	1.50

SN74LS108	3.20	3.10	2.90
SN74LS109	1.80	1.75	1.70
SN74LS110	2.80	2.75	2.65
SN74LS111	4.20	4.10	3.90
SN74LS112	1.80	1.75	1.70
SN74LS114	1.70	1.65	1.60
SN74LS116	3.20	3.10	2.95
SN74LS121	1.50	1.46	1.42

SN74LS122	1.80	1.76	1.70
SN74LS124	2.80	2.72	2.60
SN74LS128	3.70	3.60	3.42
SN74LS134	4.80	3.70	3.50
SN74LS138	4.20	4.10	3.90
SN74LS140	3.50	3.30	3.10
SN74LS141	6.50	6.30	5.90
SN74LS160	3.00	2.90	2.70
SN74LS166	2.70	2.65	2.55

SN7445	0.85	0.83	0.70
SN7446	1.03	1.00	0.85
SN7447	1.03	1.00	0.85
SN7448	0.85	0.83	0.70
SN7450	0.14	0.13	0.12
SN7451	0.14	0.13	0.12
SN7453	0.14	0.13	0.12
SN7454	0.14	0.13	0.12
SN7455	0.40	0.39	0.38
SN7460	0.14	0.13	0.12
SN7462	0.45	0.44	0.42
SN7464	0.45	0.44	0.42
SN7465	0.45	0.44	0.42
SN7470	0.30	0.27	0.25
SN7471	0.60	0.59	0.58
SN7472	0.25	0.24	0.21
SN7473	0.30	0.27	0.26
SN7474	0.31	0.29	0.26
SN7475	0.40	0.39	0.38
SN7476	0.31	0.29	0.26
SN7478	0.65	0.63	0.61
SN7480	0.43	0.41	0.36
SN7481	1.00	0.95	0.90
SN7482	0.75	0.70	0.62
SN7483	0.81	0.80	0.68
SN7484	0.90	0.86	0.85
SN7485	1.25	1.15	1.00
SN7486	0.31	0.28	0.25
SN7489	3.50	3.20	3.00
SN7490	0.45	0.42	0.35
SN7491	1.00	0.95	0.90
SN7492	0.45	0.42	0.35
SN7493	0.45	0.42	0.35
SN7494	0.48	0.45	0.40

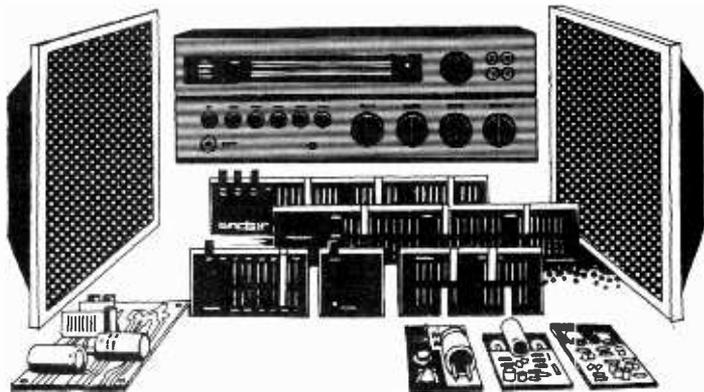
SN7495	0.60	0.56	0.50
SN7496	0.70	0.67	0.60
SN7497	0.70	0.69	0.68
SN74100	1.35	1.30	1.25
SN74104	0.31	0.29	0.26
SN74105	0.31	0.29	0.26
SN74107	0.31	0.29	0.26
SN74109	1.00	0.97	0.95
SN74110	0.55	0.50	0.45
SN74111	0.81	0.80	0.76
SN74114	1.00	0.97	0.95
SN74115	1.00	0.97	0.95
SN74118	1.00	0.95	0.90
SN74121	0.31	0.29	0.25
SN74122	0.44	0.41	0.37
SN74123	0.62	0.58	0.50
SN74125	0.70	0.65	0.60
SN74126	0.75	0.70	0.65
SN74128	1.40	1.35	1.30
SN74132	2.10	2.05	2.00
SN74136	0.95	0.90	0.85
SN74140	2.50	2.45	2.40
SN74141	0.75	0.70	0.62
SN74145	1.15	1.10	1.05
SN74147	2.95	2.90	2.85
SN74148	2.30	2.25	2.20
SN74150	1.35	1.30	1.25
SN74151	0.68	0.62	0.55
SN74152	1.55	1.50	1.45
SN74153	0.68	0.62	0.55
SN74154	1.55	1.50	1.45
SN74155	0.68	0.62	0.55
SN74156	0.68	0.62	0.55
SN74157	0.90	0.85	0.80
SN74158	1.50	1.45	1.40
SN74160	0.95	0.90	0.80
SN74161	0.95	0.90	0.80
SN74162	0.95	0.90	0.80
SN74163	0.95	0.90	0.80
SN74164	1.60	1.55	1.50

SN74165	1.60	1.55	1.50
SN74166	1.40	1.30	1.15
SN74170	2.40	2.30	2.20
SN74173	1.65	1.60	1.55
SN74174	1.15	1.10	1.00
SN74175	0.97	0.90	0.80
SN74176	1.10	1.05	1.00
SN74177	1.10	1.05	1.00
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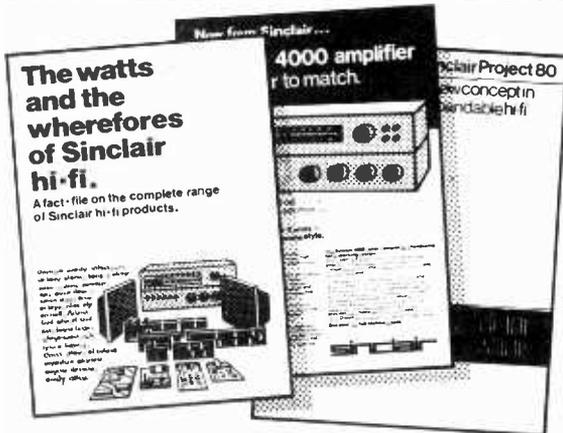
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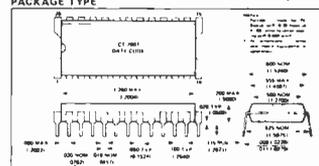
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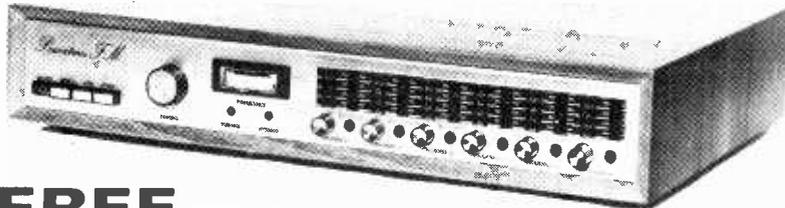
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3 Set of transistors, diodes, LED, integrated circuits for mounting on pack 1	£6.25	11 Toroidal transformer with electrostatic screen. Primary: 0-117V-234V	£4.45
4 Pre-aligned front end module, coil assembly, three-section ceramic filter	£8.80	12 Set of capacitors, rectifiers, voltage regulator for power supply	£2.95
5 Fibreglass printed circuit board for stereo decoder	£1.10	13 Set of miscellaneous parts, including sockets, fuse holder, fuses, inter-connecting wire, etc.	£1.50
6 Set of metal oxide resistors, capacitors cermet preset for decoder	£2.60	14 Set of metal work parts including silk screen printed fascia panel, acrylic silk screen printed tuning indicator panel insert, internal screen, fixing parts, etc.	£6.50
7 Set of transistors LED, integrated circuit for decoder	£3.45	15 Construction notes (free with complete kit)	£0.25
8 Set of components for channel selector switch module including fibreglass printed circuit board, push-button switches, knobs, LEDs preset adjusters, etc.	£8.30	16 Teak cabinet	£9.85
		One each of packs 1-16 inclusive are required for complete stereo FM tuner.	
		Total cost of individually purchased packs	£76.85



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NOVEL STEREO FM TUNER

In the April and May issues of *Wireless World* there was published a novel design for an fm. tuner which combines consistent high performance with the elimination of the critical setting-up procedure required by too many earlier tuners. This original circuit has been developed further and is used as the basis for our new slimline unit. The front end is a ready built pre-aligned module which then feeds an amplifier driven screened three section ceramic filter leading to an integrated circuit five-stage limiting amplifier providing excellent a.m. rejection. This is followed by a single coil integrated balanced demodulator from which the audio output may be taken. Temperature compensated varicap tuning allows stations to be selected either by a ten-turn tuning potentiometer or by a choice of six preset push-button controls. Each of the preset controls can be adjusted on the front panel with the settings being indicated by six LED lamps behind an acrylic silk screen printed fascia panel insert. Additional circuitry includes temperature compensated AFC restricted to less than station spacing, inter-station muting, a single-lamp LED tuning indicator and a linear scale frequency meter. The stereo decoder, built on a separate board, is based on a well-proven integrated circuit phase-locked-loop to which has been added active filters to remove sub-carrier harmonics and 'birdies'. The power supply, to ensure station holding stability, uses an integrated circuit voltage regulator which is powered via a low-hum field specially designed TOROIDAL TRANSFORMER.

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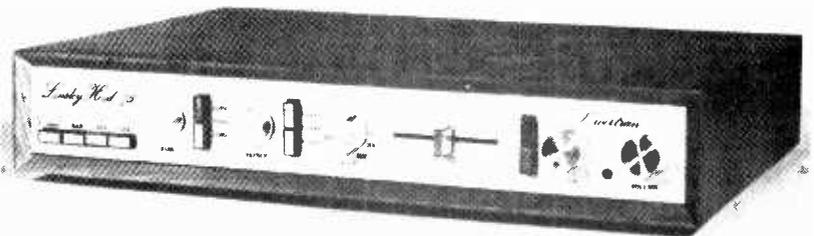
POWERTRAN ELECTRONICS

MORE ON NEXT PAGE!

DESIGNER APPROVED KIT

In *Hi-Fi News* there was published by Mr Linsley-Hood a series of four articles (November 1972-February 1973) and a subsequent follow-up article (April 1974) on a design for an amplifier of exceptional performance which has as its principal feature an ability to supply from a direct coupled fully protected output stage, power in excess of 75 watts whilst maintaining distortion at less than 0.01% even at very low power levels. The power amplifier is complemented by a pre-amplifier based on a discrete component operational amplifier referred to as the Liniac which is employed in the two most critical points of the system, namely the equalization stage and tone control stage, positions where most conventional designs run out of gain at the extremes of the frequency spectrum. Unusual features of the design are the variable transition frequencies of the tone controls and the variable slope of the scratch filter. There is a choice of four inputs, two equalized and two linear, each having independently adjustable signal level. The attractive slimline unit pictured has been made practical by highly compact PCBs and a specially designed Toroidal transformer.

Hi-Fi News Linsley-Hood 75W/Channel Amplifier Mk III Version (modifications as per *Hi-Fi News* April 1974)



Full circuit descriptor, in handbook (pack 15—price 30p)

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2 Set of resistors, capacitors, pre-sets for power amp	£1.70	12 Set of resistors, capacitors, secondary fuses semiconductors for power supply	£3.50
3 Set of semiconductors for power amp (now using BDY56, BD529, BD530)	£6.50	13 Set of miscellaneous parts including DIN skts, mains input skt, fuse holder, inter-connecting cable, control knobs	£4.25
4 Pair of 2 drilled, finned heat sinks	£0.80	14 Set of metalwork parts including silk screen printed fascia panel and all brackets, fixing parts, etc	£6.30
5 Fibreglass printed-circuit board for pre-amp.	£1.30	15 Handbook	£0.30
6 Set of low noise resistors, capacitors, pre-sets for pre-amp.	£2.70	16 Teak cabinet	£9.85
7 Set of low noise, high gain semiconductors for pre-amp.	£2.40	2 each of packs 1-7 inclusive are required for complete stereo system	
8 Set of potentiometers (including mains switch)	£2.05	Total cost of individually purchased packs	£72.25
9 Set of 4 push-button switches, rotary mode switch	£3.70		
10 Toroidal transformer complete with magnetic screen/housing primary: 0-117-234 V, secondaries: 33-0-33 V, 25-0-25 V.	£9.15		

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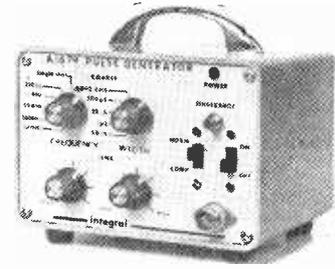
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An essential and critical component in a high-quality speaker system is the crossover unit conventionally comprising of a series of passive networks which unfortunately, though introducing reactive impedances between the amplifier and the speakers, result in the loss of the advantage of high amplifier damping factor and renders the speakers prone to overshoots and resonances. An elegant solution to this problem, described by D. C. Read in **Wireless World**, involves the use of a series of active filters splitting the output of the pre-amplifier into three channels, of closely defined bandwidth, each of which is fed to the appropriate speaker by its own power amplifier. A design for a suitable 20-watt amplifier, based on a proven Texas circuit, was also described by Mr Read. The printed-circuit board for this has been designed such that three amplifiers may be stacked and mounted together on a common heat sink to achieve a conveniently compact module.

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AC107	0.16	BC158	0.11	CV7464	0.10	2G106	0.21	SN7400	0.14	SN7413	0.30	SN7442	0.73	SN7492	0.75	No.	Qty.	Contents	Price
AC126	0.16	BC159	0.13	CV7594	0.25	2G306	0.44	SN7401	0.14	SN7416	0.30	SN7450	0.15	SN7493	0.73	J1	1	Pre-amp component kit plus data	0.65
AC127	0.16	BC171	0.16	CV7648	0.30	2G345A	0.18	SN7402	0.14	SN7417	0.29	SN7460	0.15	SN7494	0.80	J2	3	Transistors AF115 new and marked	0.65
AC128	0.12	BC172	0.16	CV8762	0.40	2G402	0.25	SN7403	0.14	SN7423	0.40	SN7472	0.30	SN7421	0.49	J4	4	Transistors 2N726 new and marked	0.65
AC138	0.20	BC173	0.16	MOS333	0.30	2N526	0.16	SN7404	0.14	SN7425	0.39	SN7473	0.41	SN7422	0.70	J5	8	Zener diodes top hat type 75 volt	0.65
AC141	0.20	BC184	0.18	MF4102	0.12	2N697	0.15	SN7405	0.15	SN7430	0.14	SN7474	0.40	SN7423	0.70	J7	50	Metres con/wire mixed colours	0.65
AC142	0.20	BC208	0.12	MJE202	0.85	2N715	0.35	SN7406	0.39	SN7432	0.39	SN7476	0.42	SN74151	1.05	J8	25	Metres con/wire 4 Metres 60/40 solder	0.65
AC153	0.22	BC209	0.13	MJE2955	0.95	2N726	0.25	SN7407	0.37	SN7433	0.40	SN7480	0.16	SN74153	0.95	J9	100	Resistors HI/STAB 1/2w mixed values	0.65
AC154	0.22	BC212L	0.14	MJE3055	0.60	2N753	0.55	SN7408	0.26	SN7440	0.15	SN7486	0.34	SN74195	1.07	J10	100	Resistors HI/STAB 1/4w mixed values	0.65
AC167	0.21	BC213L	0.12	NKT162	0.25	2N1303	0.16	SN7409	0.23	SN7441	0.70	SN7490	0.50	SN74196	1.20	J12	250	Resistors mixed values	0.65
AC168	0.27	BC214L	0.14	NKT164	0.25	2N1304	0.19	930 SERIES LOGIC I.C.s					MC948	0.25	J14	100	Capacitors miniature mixed values	0.65	
AC169	0.16	BC301	0.30	NKT212	0.20	2N1305	0.19	MC930	0.15	MC933	0.15	MC945	0.30	MC9093	0.40	J15	5	Terminal blocks brand new 12 way	0.65
AC176	0.15	BC337	0.15	NKT221	0.17	2N1309	0.25	MC932	0.15	MC944	0.16	MC946	0.14	MC9097	0.40	J15	10	Switches 5 push to make 5 off/on	0.65
AC176	0.15	MP	BC211	0.28	NKT224	0.15	2N1754	0.20						J18	12	Standard crocodile clips	0.65		
AC128	0.25	BD131	0.40	NKT270	0.15	2N2484	0.30						J19	12	Screwdrivers 5 inches long	0.65			
AC177	0.26	BD132	0.40	NKT278	0.15	2N2923	0.16						J20	1	Pack nuts & bolts self tappers, etc.	0.65			
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2924	0.18						J22	20	Volume controls lin & log mixed values	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2925	0.15						J23	75	Syn/rubber grommets mixed sizes	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2926	0.14						J25	20	Screw on rubber feet 1/4 inch dia approx	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2927	0.12						J26	1	Pack marker sleeve mixed colours & sizes	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2928	0.11						J27	5	lengths ferrite rod mixed flat & round	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2929	0.15						J28	20	Tag strips assorted lengths	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2930	0.14						J29	4	Micro switches brand new	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2931	0.15						J31	20	Preset pots lin & log mixed	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2932	0.14						J32	20	Capacitors can type mixed values	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2933	0.15						J33	50	Ceramic plate capacitors mixed values	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2934	0.14						J34	1	Pack copper clad pavoin board	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2935	0.15						J35	20	Fuse holders mixed	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2936	0.14						J36	8	Metres multi/core solder 60/40 22swg	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2937	0.15						J37	10	3.5mm Jack sockets chrome	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2938	0.14						J38	8	Sockets 5 pin & 2 pin	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2939	0.15						J39	2	Relays 12 & 24 volt	0.65		
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2940	0.14						J40	1	Pack of component boards ICs Transistors etc	0.65		
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2941	0.15						J PACKS ONLY ADD 10% TO TOTAL ORDER PLEASE FOR POST & PACK					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2942	0.14						PAPST TAPE MOTOR 220v 50Hz £2.50 p&p 25p					
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AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2947	0.15						RELAY T.M.C. MINIATURE 3 300 ohms 2 pole 4 way 0.55					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2948	0.14						SWITCH 4 POLE 2 CHANGE/OVER 0.15					
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2949	0.15						OUR BONANZA CAPACITORS PACK IS STILL AVAILABLE ONLY £1.50					
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AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2951	0.15						TWIN V/U METERS PLASTIC APPROX 40mm By 80mm BRAND NEW READS 20 TO 1 MINUS ZERO TO 3 PLUS. L/R BARGAIN ONLY £2.95					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2952	0.14						MINIATURE EDGE LEVEL METER LOTS OF USES ONLY 0.95					
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AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2954	0.14						PLEASE ADD 10p MIN. TO ALL ORDERS WHERE NOT STATED FOR P&P					
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2955	0.15						VAT INCLUDED AT CURRENT RATE OVERSEAS POST AT COST					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2956	0.14						MAIL ORDER DEPT. ONLY					
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2957	0.15						(Callers by appointment)					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2958	0.14						J.E.T. ELECTRONICS					
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2959	0.15						90A MAWNEY ROAD, ROMFORD, ESSEX RM7 7DA					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2960	0.14						TELEPHONE: ROMFORD 61486					
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2961	0.15						SPECIAL OFFERS ARE AVAILABLE ONLY WHILE STOCK LASTS					
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2962	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2963	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2964	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2965	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2966	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2967	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2968	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2969	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2970	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2971	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2972	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2973	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2974	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2975	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2976	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2977	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2978	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2979	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2980	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2981	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2982	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2983	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2984	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2985	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2986	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2987	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2988	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2989	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2990	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2991	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2992	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2993	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2994	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2995	0.15											
AC178	0.25	BD131	0.40	MP	OC22	0.50	2N2996	0.14											
AC178	0.25	BD132	0.40	MP	OC28	0.50	2N2997	0.15											
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AC178	0.25	BD132	0.40	MP	OC28	0.50	2N3001	0.15											
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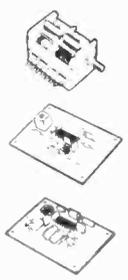
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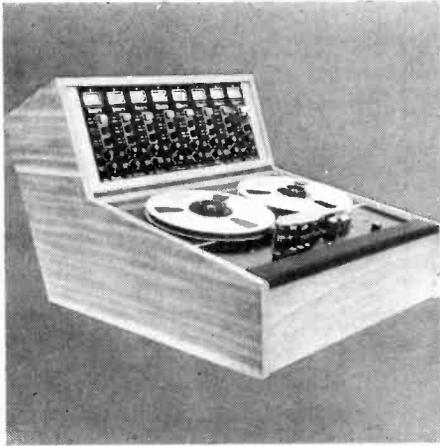
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PLASTIC SNAP PACK	BASF LH		BASF SUPER S/M		BASF Cr 02		SCOTCH DYNARANGE		SCOTCH Cr 02	
	ONE	10	ONE	10	ONE	10	ONE	10	ONE	10
C60	£0.44	£4.35	£0.57	£5.60	£0.88	£8.70	£0.46	£4.50	£0.88	£8.70
C90	£0.84	£6.20	£0.79	£7.30	£1.16	£11.55	£0.65	£6.45	£1.18	£11.70
C120	£0.80	£7.90	£1.06	£10.50	£1.47	£14.65	£0.87	£8.60	£1.47	£14.65

PLASTIC SNAP PACK	AGFA LH		AGFA G02		MEMOREX MRX		MEMOREX CRO2		TDK DYNAMIC	
	ONE	10	ONE	10	ONE	10	ONE	10	ONE	10
C60	£0.38	£3.70	£0.71	£6.90	£0.61	£6.00	£0.84	£8.30	£0.43	£4.25
C90	£0.49	£4.86	£0.94	£9.35	£0.87	£8.60	£1.11	£11.00	£0.68	£6.35
C120	£0.68	£6.50	£1.27	£12.80	£1.17	£11.20	—	—	£0.88	£8.80

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	ONE	10	ONE	10	ONE	10	ONE	10
LONG PLAY								
5" 900'	£1.55	£15.00	£1.84	£17.90	—	£12.90	£1.22	£12.10
5 1/4" 1200'	£1.78	£17.50	£2.40	£22.76	—	—	£1.60	£15.00
7" 1800'	£2.32	£23.00	£2.60	£25.50	£2.10	£20.50	£2.18	£21.00
DOUBLE PLAY								
5" 1200'	£1.80	£17.75	£2.40	£22.76	£1.48	£14.00	£1.52	£14.49
5 1/4" 1800'	£2.57	£25.00	£3.00	£28.52	—	—	£2.20	£21.00
7" 2400'	£2.85	£28.00	£3.36	£32.50	£2.60	£25.00	£2.60	£25.50
TRIPLE PLAY								
5" 1800'	£2.52	£25.00	£3.00	£28.52	—	—	£2.30	£22.20
5 1/4" 2400'	£3.20	£31.00	£3.85	£35.95	—	—	£2.85	£27.75
7" 3600'	£3.90	£38.00	£5.25	£51.92	—	—	£3.50	£34.50
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Wireless system components.

Varicap tuning accessories:

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EC720: Ant, RF and Osc. tuned by MVAM varicap diode. IC signal processing system, ceramic IF filter. For ferrite rod or loop antenna. Kit ... £8.00

MVAM1 3x300pF varicap diode. £2.75, or MVAM2 2 diode version £1.05. (2% match)

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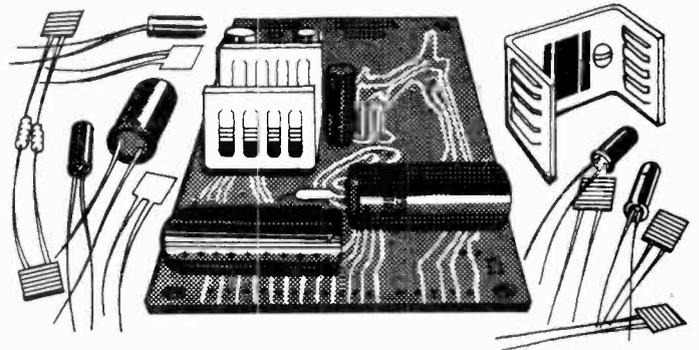
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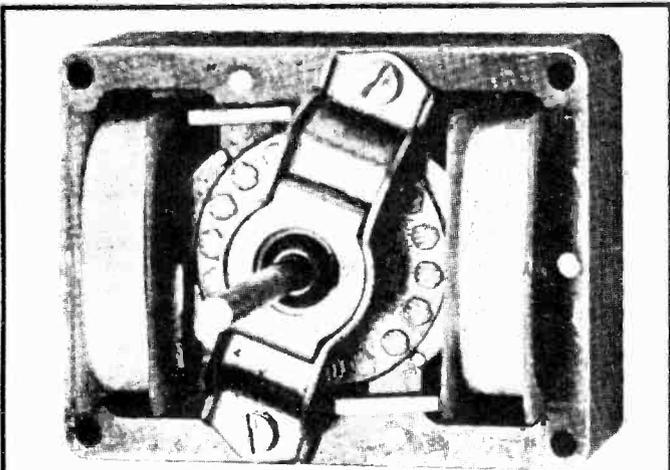
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AC126 0.17	BC142 0.23	BCY70 0.14	BF123 0.28	BLV15A 2.93	BU126 1.61	BZX61-C72 0.23	OC171 0.23	1N4740A 0.19	2N3439 0.69
AC127 0.18	BC143 0.23	BCY71 0.21	BF125 0.25	BRC4443 0.86	BU206 2.30	BZX70-C30 0.34	OC172 0.23	1N4742A 0.19	2N3440 0.83
AC128 0.17	BC144 0.25	BCY72 0.14	BF127 0.28	BSV64 0.86	BU205 2.41	BZX83 Series 0.11	OC200 0.63	1N4746A 0.19	2N3442 0.92
AC141 0.18	BC147 0.11	BD115 0.63	BF157 0.23	BSV68 0.46	BU206 2.76		OC206 0.23	1N4749A 0.19	2N3702 0.13
AC142 0.21	BC148 0.11	BD130 0.80	BF177 0.28	BSV68 0.46	BU208 3.56		OC207 1.43	1N4751A 0.19	2N3703 0.13
AC141K 0.32	BC149 0.11	BD131 0.36	BF178 0.29	BSX20 0.17	BY100 0.17		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC142K 0.30	BC152 0.17	BD132 0.50	BF179 0.34	BSX21 0.23	BY100 0.17		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC176 0.32	BC153 0.19	BD133 0.66	BF181 0.34	BSX76 0.21	BY103 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC187 0.21	BC157 0.13	BD135 0.66	BF185 0.19	BSY41 0.46	BY133 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC187K 0.34	BC158 0.13	BD136 0.39	BF194 0.11	BSY52 0.41	BY142 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC188 0.21	BC159 0.13	BD137 0.41	BF195 0.12	BSY53 0.39	BY184 0.32		OC207 1.11	1N4751A 0.19	2N3704 0.13
AC188K 0.32	BC160 0.32	BD138 0.43	BF196 0.14	BSY54 0.44	BY198 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AD140 0.48	BC161 0.34	BD139 0.47	BF197 0.14	BSY55 0.74	BY199-400 0.21		OC207 1.11	1N4751A 0.19	2N3704 0.13
AD142 0.52	BC168B 0.11	BD140 0.51	BF219 0.28	BSY56 0.82	BY201-2 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AD143 0.46	BC171A 0.11	BD144 2.53	BF220 0.28	BSY65 0.16	BY201-3 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
AD149 0.55	BC172 0.11	BD181 0.86	BF224J 0.14	BSY76 0.23	BY201-6 0.40		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF114 0.14	BC182 0.14	BD182 0.92	BF240 0.34	BSY78 0.25	BY203-12 0.17		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF115 0.14	BC182L 0.14	BD183 0.97	BF244 0.17	BSY95A 0.11	BY203-16 0.24		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF116 0.14	BC183 0.10	BD226 0.55	BF245A 0.34	BT101-300R 1.81	BY204-4 0.27		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF117 0.14	BC183L 0.10	BD227 -0.55	BF257 0.34	BT101-500R 1.72	BY204-8 0.29		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF118 0.52	BC184 0.13	BD228 0.82	BF258 0.41	BT102-300R 1.38	BY204-10 0.34		OC207 1.11	1N4751A 0.19	2N3704 0.13
AF367 0.63	BC184L 0.13	BD229 0.82	BF259 0.57	BT102-500R 1.49	BY204-12 0.36		OC207 1.11	1N4751A 0.19	2N3704 0.13
AFZ11 1.10	BC212 0.11	BD230 0.66	BF336 0.34	BT106 0.97	BY205-1 0.17		OC207 1.11	1N4751A 0.19	2N3704 0.13
AFZ12 1.32	BC212L 0.16	BD231 0.66	BF337 0.40	BT107 1.72	BY205-12 0.17		OC207 1.11	1N4751A 0.19	2N3704 0.13
AL100 0.78	BC213 0.14	BD233 0.52	BF338 0.44	BT108 1.72	BY220-200 0.23		OC207 1.11	1N4751A 0.19	2N3704 0.13
AL102 0.74	BC213L 0.14	BD234 0.54	BF341 0.40	BT109 1.12	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
AL103 0.74	BC214 0.16	BD235 0.58	BF342 0.40	BT116 0.97	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA102 0.18	BC214L 0.16	BD236 0.81	BF343 0.36	BT119 2.53	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA108 0.41	BC237 0.17	BD237 0.84	BFW30 1.97	BT120 2.53	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA115 0.08	BC238 0.17	BD238 0.87	BFW59 0.18	BTX18-100 0.96	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA130 0.14	BC300 0.34	BD433 0.74	BFW60 0.18	BTX18-200 1.15	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA141 0.31	BC301 0.34	BD434 0.78	BFX29 0.29	BTX18-400 1.61	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA144 0.16	BC303 0.46	BD435 0.86	BFX30 0.29	BTX18-500 1.72	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA145 0.15	BC307 0.11	BD436 0.94	BFX34 0.97	BTY79-100R 2.71	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA148 0.14	BC327 0.17	BD437 0.96	BFX52 0.43	BTY79-200R 3.08	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA154 0.14	BC329 0.17	BD438 1.10	BFX84 0.25	BTY79-300R 3.33	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA155 0.13	BC337 0.17	BD439 1.10	BFX85 0.29	BTY79-400R 3.80	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BA156 0.14	BC338 0.17	BDY10 3.04	BFX86 0.23	BTY79-500R 4.16	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC107 0.16	BC377 0.23	BDY11 1.03	BFX87 0.23	BTY79-600R 4.53	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC108 0.15	BCY30 0.49	BDY20 0.92	BFX88 0.25	BTY79-800R 6.67	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC109 0.16	BCY31 0.57	BDY38 0.89	BFY18 0.34	BTY87-100R 3.22	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC107B 0.18	BCY32 1.23	BDY60 0.80	BFY40 0.55	BTY87-200R 3.56	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC108C 0.18	BCY33 0.40	BDY61 0.89	BFY41 0.57	BTY87-400R 4.14	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC109C 0.23	BCY34 0.52	BDY62 0.86	BFY50 0.23	BTY87-600R 5.06	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC117 0.21	BCY38 0.48	BDY90 3.27	BFY51 0.21	BTY91-200R 4.51	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC125 0.16	BCY39 1.43	BDY91 3.13	BFY52 0.22	BTY91-400R 5.29	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13
BC126 0.21	BCY40 0.78	BDY92 2.53	BFY53 0.23	BTY91-600R 6.39	BY220-800 0.28		OC207 1.11	1N4751A 0.19	2N3704 0.13

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50 V	0.29	0.40	0.41	0.42	0.47	LIT704 1.03
100 V	0.29	0.46	0.47	0.48	0.54	LIT707 1.03
200 V	0.38	0.58	0.58	0.58	0.68	LIT747 2.01
400 V	0.46	0.70	0.87	0.88	0.98	
600 V	0.58	0.86	1.09	1.19	1.26	

TRIACS	1.6 amp T05	4 amp T0220	6.5 amp T0220	8.5 amp T0220	10 amp T0220	16 amp T0220
100 V	0.32 0.32	0.60 0.60	0.70 0.70	0.78 0.78	0.83 0.83	1.01 1.01
200 V	0.34 0.34	0.64 0.64	0.75 0.75	0.87 0.87	0.87 0.87	1.17 1.17
400 V	0.41 0.44	0.77 0.78	0.80 0.83	0.97 1.01	1.13 1.19	1.70 1.74
600 V	0.51 0.58	0.96 0.99	0.87 1.01	1.21 1.26	1.42 1.50	2.11 2.17

N.B. TRIACS ARE AVAILABLE WITH OR WITHOUT AN INTERNAL TRIGGER DIAC (a) INDICATES PRICES FOR TRIACS WITHOUT AN INTERNAL DIAC (b) INDICATES PRICES OF TRIACS WITH AN INTERNAL TRIGGER DIAC PLEASE INDICATE CLEARLY WHICH TYPE OF DEVICE IS REQUIRED

BZK61-C51 0.23	OC140 0.34	1N4149 0.07	2N3233 1.09
BZK61-C68 0.23	OC170 0.23	1N4454 0.09	2N3415 0.14
BZK61-C72 0.23	OC171 0.23	1N4740A 0.19	2N3439 0.69
BZX70-C30 0.34	OC172 0.23	1N4742A 0.19	2N3440 0.83
BZX83 Series 0.11	OC200 0.63	1N4746A 0.19	2N3442 0.92
BZY88 Series 0.11	OC206 1.43	1N4749A 0.19	2N3702 0.13
C106F 0.40	OC207 1.11	1N4751A 0.19	2N3703 0.13
C106A 0.46	OC207 1.11	1N4751A 0.19	2N3704 0.13
C106B 0.57	OC207 1.11	1N4751A 0.19	2N3704 0.13
C106D 0.69	OC207 1.11	1N4751A 0.19	2N3704 0.13
C111 0.28	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS1/05 0.28	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS1/10 0.28	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS1/20 0.38	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS1/40 0.46	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS3/05 0.34	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS3/10 0.46	OC207 1.11	1N4751A 0.19	2N3704 0.13
CRS3/20 0.52	OC207 1.11	1N4751A 0.19	2N3704 0.13
TBA641 0.95	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIC44 0.36	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIC45 0.39	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIC46 0.49	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIC47 0.69	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIL209 0.18	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP29A 0.49	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP30A 0.58	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP31A 0.62	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP32A 0.71	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP41A 0.74	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIP42A 0.80	OC207 1.11	1N4751A 0.19	2N3704 0.13
TIS26 1.94	OC207 1.11	1N4751A 0.19	2N3704 0.13
UJ235 2.53	OC207 1.11	1N4751A 0.19	2N3704 0.13
ZFX109 0.15	OC207 1.11	1N4751A 0.19	2N3704 0.13
ZTX313 0.25	OC207 1.11	1N4751A 0.19	2N3704 0.13
ZTX504 0.44	OC207 1.11	1N4751A 0.19	2N3704 0.13
1N542 0.07	OC207 1.11	1N4751A 0.19	2N3704 0.13
1N645 0.23	OC207 1.11	1N4751A 0.19	2N3704 0.13
1N746A 0.14	OC207 1.11	1N4751A 0.19	2N3704 0.13
1N750A 0.14	OC207 1.11	1N4751A 0.19	2N3704 0.13
1N711 0.21	OC207 1.11	1N4751A 0.19	2N3704 0.13
2N2102 0.52	OC207 1.11	1N4751A 0.19	2N3704 0.13
2N2117 0.34	OC207 1.11	1N4751A 0.19	2N3704 0.13
2N2219A 0.52	OC207 1.11	1N4751A 0.19	2N3704 0.13
2N2221A 0.40	OC207 1.11	1N4751A 0.19	2N3704 0.13

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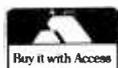
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ASV51	0.26	BF152	0.56
ASV52	0.26	BF153	0.46
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ASV55	0.26	BF155	0.46
ASV56	0.26	BF156	0.46
ASV57	0.26	BF157	0.46
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ASV73	0.26	BF159	0.61
ASZ21	0.41	BF160	0.41
BC107	0.08	BF161	0.41
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BC109	0.08	BF164	0.41
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BC114	0.16	BF167	0.22
BC115	0.16	BF173	0.22
BC116	0.16	BF176	0.36
BC117	0.19	BF177	0.31
BC118	0.10	BF178	0.31
BC119	0.31	BF179	0.31
BC120	0.81	BF180	0.31
BC125	0.12	BF181	0.31
BC126	0.19	BF182	0.41
BC132	0.12	BF183	0.41
BC134	0.19	BF184	0.26
BC135	0.12	BF185	0.31
BC136	0.16	BF187	0.28
BC137	0.16	BF188	0.41
BC139	0.31	BF194	0.12
BC140	0.31	BF267	0.12
BC141	0.31	BF196	0.15
BC143	0.31	BF197	0.15
BC145	0.46	BF200	0.46
BC147	0.10	BF222	0.28
BC148	0.10	BF225	0.61
BC149	0.19	BF259	0.87
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BC152	0.18	BF263	0.56
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7402	0.14 0.13 0.12	7490	0.60 0.58 0.56
7403	0.14 0.13 0.12	7491	1.02 0.97 0.93
7404	0.14 0.13 0.12	7492	0.69 0.66 0.59
7405	0.14 0.13 0.12	7493	0.69 0.66 0.59
7406	0.36 0.31 0.29	7494	0.79 0.76 0.69
7407	0.36 0.31 0.29	7495	0.79 0.76 0.69
7408	0.23 0.22 0.21	7596	0.89 0.86 0.80
7409	0.28 0.27 0.26	74100	1.39 1.34 1.30
7410	0.14 0.13 0.12	74104	0.56 0.54 0.51
7411	0.23 0.22 0.21	74105	0.56 0.54 0.51
7412	0.26 0.25 0.24	74107	0.41 0.39 0.37
7413	0.30 0.29 0.28	74110	0.56 0.51 0.46
7416	0.28 0.27 0.26	74111	0.83 0.81 0.78
7417	0.28 0.27 0.26	74118	0.93 0.88 0.83
7420	0.14 0.13 0.12	74119	1.39 1.30 1.20
7422	0.28 0.27 0.26	74121	0.46 0.44 0.41
7423	0.37 0.36 0.35	74122	0.65 0.63 0.60
7425	0.37 0.36 0.35	74123	0.69 0.68 0.65
7426	0.37 0.35 0.33	74141	0.79 0.76 0.73
7427	0.37 0.35 0.33	74145	1.20 1.16 1.11
7428	0.42 0.39 0.37	74150	1.39 1.30 1.20
7430	0.14 0.13 0.12	74151	1.02 0.97 0.93
7432	0.37 0.35 0.33	74153	0.93 0.88 0.83
7433	0.39 0.37 0.35	74154	1.57 1.43 1.48
7437	0.32 0.30 0.28	74155	1.11 1.06 1.02
7438	0.32 0.30 0.28	74156	1.67 1.62 1.55
7440	0.14 0.13 0.12	74157	1.67 1.62 1.55
7441	0.69 0.66 0.59	74160	1.48 1.44 1.39
7442	0.69 0.66 0.59	74161	1.48 1.44 1.39
7443	1.11 1.06 1.02	74162	1.30 1.25 1.20
7444	1.11 1.06 1.02	74163	1.30 1.25 1.20
7445	1.48 1.44 1.39	74164	1.67 1.62 1.55
7446	1.11 1.06 1.02	74165	1.67 1.62 1.55
7447	1.02 0.99 0.97	74166	1.48 1.44 1.39
7448	1.02 0.99 0.97	74174	1.48 1.44 1.39
7450	0.14 0.13 0.12	74175	1.02 0.97 0.93
7451	0.14 0.13 0.12	74176	1.16 1.11 1.06
7453	0.14 0.13 0.12	74177	1.16 1.11 1.06
7454	0.14 0.13 0.12	74180	1.16 1.11 1.06
7460	0.14 0.13 0.12	74181	3.66 3.56 3.47
7470	0.30 0.27 0.25	74182	1.16 1.11 1.06
7472	0.30 0.27 0.25	74184	1.67 1.62 1.55
7473	0.38 0.36 0.32	74190	1.81 1.76 1.71
7474	0.38 0.36 0.32	74191	1.81 1.76 1.71
7475	0.58 0.56 0.52	74192	1.81 1.76 1.71
7476	0.11 0.10 0.09	74193	1.81 1.76 1.71
7480	0.56 0.54 0.51	74194	1.02 1.16 1.11
7481	1.02 0.97 0.93	74195	1.02 0.97 0.93
7482	0.83 0.79 0.74	74196	1.11 1.06 1.02
7483	1.11 1.06 0.97	74197	1.11 1.06 1.02
7484	0.98 0.88 0.88	74198	2.55 2.50 2.45
7485	1.48 1.44 1.39	74199	2.31 2.21 2.11

Devices may be mixed to qualify for quantity price (TTL 74 series only) data is available for the above series of I.C.'s in booklet form. PRICE 35p

* D.T.L. 930 SERIES

Type	Quantities	Type	Quantities
	1 25 100+		25 100+
BP930	0.14 0.13 0.12	BP948	0.28 0.26 0.23
BP932	0.15 0.14 0.13	BP951	0.65 0.60 0.56
BP933	0.15 0.14 0.13	BP952	0.14 0.13 0.12
BP935	0.15 0.14 0.13	BP953	0.42 0.40 0.38
BP936	0.15 0.14 0.13	BP954	0.42 0.40 0.38
BP944	0.15 0.14 0.13	BP957	0.42 0.40 0.38
BP945	0.28 0.26 0.23	BP959	0.42 0.40 0.38
BP946	0.14 0.13 0.12		

Devices may be mixed to qualify for quantity price. Larger quantity prices on application: (D.T.L. 930 Series only).

* THYRISTORS

PIV	0.6A	0.8A	1A	3A	5A	7A	10A	16A	30A
	TO18	TO18	TO5	TO66	TO64	TO48	TO48	TO48	TO48
10	0.13 0.15	—	—	—	—	—	—	—	—
20	0.15 0.18	—	—	—	—	—	—	—	—
30	0.19 0.22	—	—	—	—	—	—	—	—
50	0.22 0.28	0.20 0.25 0.36	0.36 0.48 0.51	0.54 1.18	1.18	—	—	—	—
100	0.25 0.30	0.25 0.25 0.48	0.48 0.51 0.57	0.58 1.43	1.43	—	—	—	—
150	0.31 0.38	—	—	—	—	—	—	—	—
200	0.38 0.44	0.25 0.30 0.50	0.50 0.57 0.62	0.62 1.63	1.63	—	—	—	—
300	—	0.30 0.39 0.55	0.57 0.62 0.71	0.71 1.79	1.79	—	—	—	—
600	—	0.39 0.48 0.69	0.69 0.78 0.99	0.99 2.00	2.00	—	—	—	—
800	—	0.58 0.65 0.81	0.81 0.92 1.22	1.39 4.07	4.07	—	—	—	—

* LINEAR I.C.'s

Type	Quantities	Type	Quantities
	1 25 100+		1 25 100+
72702	0.46 0.41 0.42	IAA350A	1.71 1.67 1.57
72709	0.23 0.21 0.19	uA703C	0.26 0.24 0.22
72709P	0.19 0.18 0.17	uA709C	0.19 0.18 0.17
72710	0.32 0.31 0.28	uA711C	0.32 0.31 0.28
72711	0.28 0.27 0.26	uA712C	0.32 0.31 0.28
72711C	0.26 0.25 0.24	uA723C	0.45 0.43 0.40
72711P	0.28 0.27 0.26	76003	1.39 1.34 1.30
72717	0.79 0.74 0.61	76023	1.39 1.34 1.30
72748P	0.35 0.33 0.31	76660	0.88 0.86 0.83
SI201C	0.46 0.42 0.37	LM380	0.83 0.80 0.83
SL701C	0.46 0.42 0.37	NE555	0.45 0.43 0.40
SL702C	0.46 0.42 0.37	NE556	

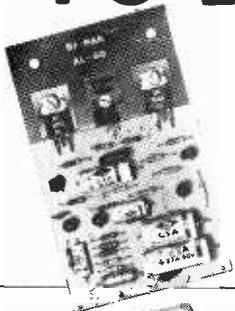
PO BOX 6 WARE HERTS

AL 60

ONLY £3.95

50w. PEAK (25w. R.M.S.)

- Max Heat Sink temp 90°C
- Frequency Response 20Hz to 100K Hz
- Distortion better than 0.1 at 1KHz
- Supply voltage 15-50 volts
- Thermal Feedback
- Latest Design Improvements
- Load — 3, 4, 5 or 16 ohms
- Signal to noise ratio 80dB
- Overall size 63mm x 105mm x 13mm. Especially designed to a strict specification. Only the finest components have been used and the latest solid state circuitry incorporated in this powerful little amplifier which should satisfy the most critical A.F. enthusiast.



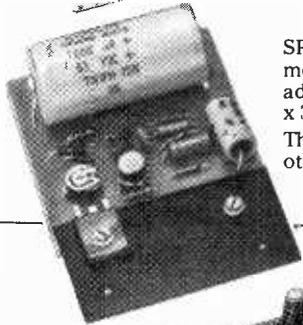
STABILISED POWER MODULE SPM80

SPM80 is especially designed to power 2 of the AL60 Amplifiers, up to 15 watt (r.m.s.) per channel simultaneously. This module embodies the latest components and circuit techniques incorporating complete short circuit protection. With the addition of the Mains Transformer BMT80, the unit will provide outputs of up to 1.5 amps at 35 volts. Size: 63mm x 105mm x 30mm.

These units enable you to build Audio Systems of the highest quality at a hitherto unobtainable price. Also ideal for many other applications including:—Disco Systems. Public Address Intercom Units, etc. Handbook available 10p.

TRANSFORMER BMT80 £2.60

PRICE £3.00



STEREO PRE-AMPLIFIER TYPE PA100

Built to a specification and NOT a price, and yet still the greatest value on the market, the PA100 stereo pre-amplifier has been conceived from the latest circuit techniques. Designed for use with the AL50 power amplifier system, this quality made unit incorporates no less than eight silicon planar transistors, two of these are specially selected low noise NPN devices for use in the input stages.

Three switched stereo inputs, and rumble and scratch filters are features of the PA100 which also has a STEREO/MONO switch, volume, balance and continuously variable bass and treble controls.

£13.20

MK 60 AUDIO KIT

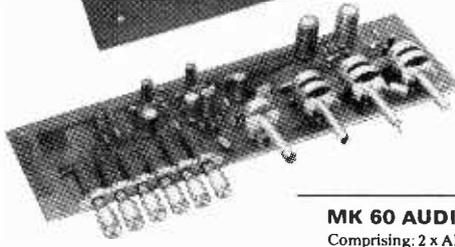
Comprising: 2 x AL60, 1 x SPM80, 1 x BTM80, 1 x PA100, 1 front panel, 1 kit of parts to include on-off switch, neon indicator, stereo headphone sockets plus instruction booklets.

COMPLETE PRICE: £27.55 plus 45p postage.

TEAK 60 AUDIO KIT

Comprising: Teak veneered cabinet size 16 1/4" x 11 1/4" x 3 3/4", other parts include aluminium chassis, heatsink and front panel bracket, plus back panel and appropriate sockets, etc.

KIT PRICE: £9.20 plus 45p postage.



STEREO 30 COMPLETE AUDIO CHASSIS

7 + 7 WATTS R.M.S.

The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supply. This with only the addition of a transformer or overwind, will produce a high quality audio unit suitable for use with a wide range of inputs, i.e. high quality ceramic pickup, stereo tuner, stereo tape deck, etc.

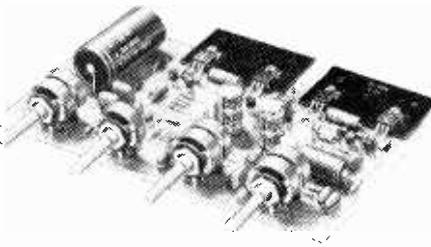
Simple to install, capable of producing really first-class results, this unit is supplied with full instructions, black front panel, knobs, mains switch, fuse & fuse holder and universal mounting bracket, enabling it to be installed in a record plinth, cabinets of your own construction or the cabinet available.

Ideal for the beginner or advanced constructor who requires Hi-Fi performance with a minimum of installation difficulty. Can be installed in 30 mins.

PRICE £15.75 Plus 45p postage & packing

TRANSFORMER £2.45 plus 45p postage & packing

TEAK CASE £3.65 plus 45p postage & packing



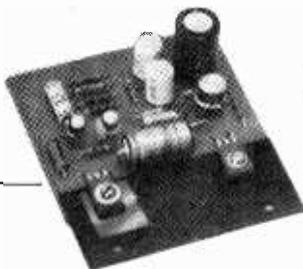
PLEASE ADD V.A.T. AT 25% TO ALL ITEMS EXCEPT
 * ADD 8%
 # NO V.A.T.
 GIRO NUMBER 388-7006

AL 10/AL 20/AL 30

The AL10, AL20 and AL30 units are similar in their appearance and in their general specification. However, careful selection of the plastic power devices has resulted in a range of output powers from 3 to 10 watts R.M.S.

The versatility of their design makes them ideal for use in record players, tape recorders, stereo amplifiers and cassette and cartridge tape players in the car and at home.

AL10 £2.30, AL20 £2.65, AL30 £2.95



SPEAKERS

E.M.I. LEK 350 Loudspeakers Enclosure kit in teak veneer, including speakers. Rec. retail price £4.50 per pair.
OUR SPECIAL PRICE ONLY £27.75 per pair P&P £3
WHILE STOCKS LAST!

HEADPHONES

4-16 ohms impedance frequency response 20 to 20,000 Hz stereo/mono switch and Volume Control £4.55

FRONT PANEL

FOR PA100. Attractive matt silver. Finish with black trim and lettering. Adds that professional touch. **£1.10 only.**

M.P.A.30

Enjoy the quality of a magnetic cartridge with your existing ceramic equipment using the new 3i-Pak M.P.A.30 which is a high quality pre-amplifier exist for the use of ceramic cartridges only.

Used in the construction are 4 low noise, high gain, silicon transistors and it is provided with a standard DIN input socket for ease of connection.

Supplied with full, easy to follow instructions.

PRICE £2.65

STORAGE-CARRY CASES

- RECORD CASES**
 7 in E.P. 18 3/8th in. x 7 in x 8 in (50 records) ***£2.48**
 12 in L.P. 13 3/4 in x 7 3/8th in x 1 1/2 in (50 records) ***£3.30**
- CASSETTE CASES**
 Holds 15. 10 in x 3 3/4 in x 5 in. Lock and handle ***£1.50**
- 8-TRACK CARTRIDGE CASES**
 Holds 14. 13 in x 5 in x 6 in. Lock and handle ***£2.20**
 Holds 24. 13 3/8th in x 8 in x 5 3/8th in Lock and handle ***£3.20**

CARTRIDGES

- ACOS GP91-ISC200mV at 1.2cm/s/sec **£1.11**
 GP93-1 280mV at 1cm/sec **£1.43**
 GP96-1 100mV at 1cm/sec **£2.31**
 TTC J-2005 Crystal/Hi Output **£0.97**
 J-2010C Crystal/Hi Output **£1.11**
 Compatible **£1.52**
 J-2006S Stereo/Hi Output **£1.81**
 J-2105 Ceramic/Med Output **£1.11**
 J-2203 Magnetic 5mV/5cm/sec including stylus **£4.78**
 J-22038 Replacement stylus for above **£2.88**
 AT-55 Audio-technica magnetic cartridge 4mV/5cm/sec **£3.06**

DYNAMIC MICROPHONE

TYPE B1223 200 ohms impedance. Complete with stand, on/off switch and 2.5mm and 3.5mm plugs. Suitable for cassette tape recorders. **PRICE £1.07**

JUST OUT!

STEREO FM TUNER

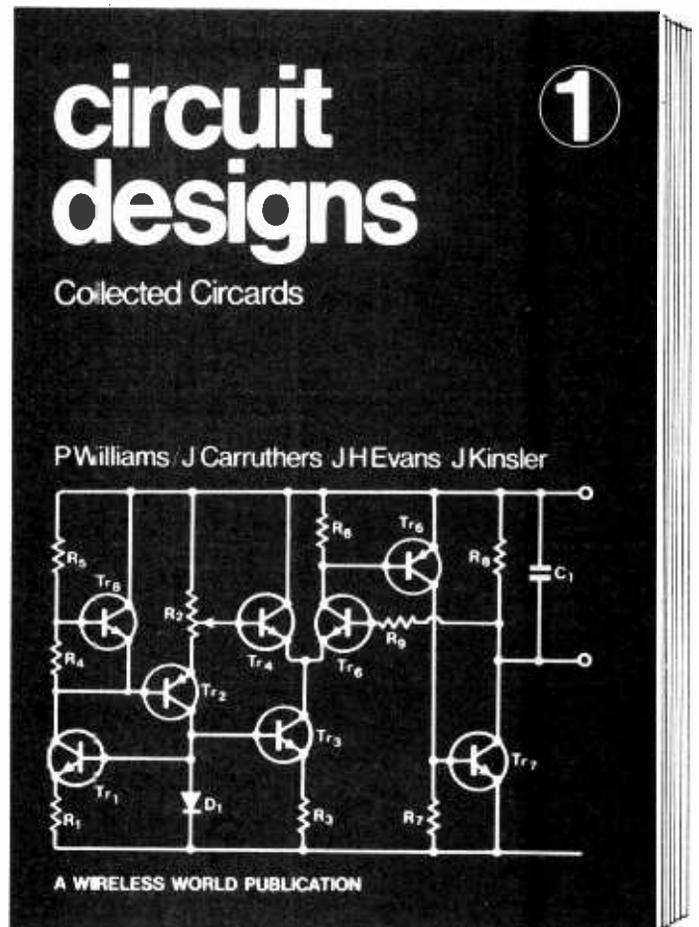
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For all who want to know about electronic circuits

Here's a book of very special appeal to all concerned with designing, using or understanding electronic circuits. It comprises information previously included in the first ten sets of Wireless World's highly successful Circards – regularly published cards giving *selected* and *tested* circuits, descriptions of circuit operation, component values and ranges, circuit limitations, modifications, performance data and graphs. Each of the ten sets – including additional circuits – in this magazine size hard cover book has been updated where necessary, and is preceded by an explanatory introduction. Circuit designs (1) is the first collection of its kind.

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- Constant-current circuits**
- Power amplifiers**
- Astable circuits**
- Optoelectronics**
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PERMEABILITY TUNERS

MW two stage ideal for use with ZN414 or similar circuit. Price 15p each + post and VAT 15p



EDGE MOUNTING MOVING COIL METER

Size 2 1/4" x 1" by Weston. 100 uA movement scaled DB, unused still in original maker's cartons. £2.50 each + post and VAT 50p.



HIGHLY SENSITIVE MOVING COIL RELAY

panel mounting with glass window this measures approx. 5 1/2" x 4" x 5" triggering current can be varied from a fraction of a milliamp to 5 milliamps by removing the front and adjusting the setting level. Price £8 each + post and VAT 95p



DC HIGH CURRENT PANEL METERS

3 1/2" wound wide angle 240 movement meters, flush mounting fitted with external shunts, made by Compton Parkinson brand new, still in maker's cartons. These are a real bargain at £5.50 each. Reasonable quantities available in the following ranges: 0-10 amps, 0-20 amps, 0-30 amps, 0-40 amps, 0-50 amps. Post and VAT 80p each



24v POWER PACK

Normal mains input with a thermal safety device, 800 mA output. 4000 mfd of smoothing and full wave rectification, completely enclosed in plastic box and with flex for mains and terminal block for output. Price £1.75 + £1.10 post and VAT



GPO PUSH BUTTON DIALLING UNIT

Will take the place of the normal rotating dial, has 10 numbered keys, so suitable for other digital systems. A desk mounting unit with rubber feet, this is a very intricate and expensive piece of apparatus. New and unused — our price only £9 each + £1.36 post and VAT



OVEN THERMOSTAT

Made by the famous Diamond H Company, this has a sensor joined by a capillary to a variable control and when fitted with a knob is ideal for many ovens or processes. 50p each + post and VAT 15p



AUDIO AMPLIFIER

Part of the famous Reditude background music system, secondhand, but believed in good order. However, no guarantee; we are selling for spares value only. These are 6 valve amplifiers, the output valves are 2 x EL 84 in push/pull, complete with mains transformer, rectifier and ample smoothing equipment. The mains transformer alone, today, would cost at least £4. Size is 9 1/2" x 5 1/2" x 4 1/2". Price only £2.00 + postage and VAT £1.50



BREAK-DOWN UNIT

Contains hundreds of useful parts some of which are as follows — 66 silicon diodes equivalent OA 91, 68 resistors, mostly 1/2 watt 5% covering a wide range of values, 4 x 1 mfd 400V oil condensers, 15 x .01 mfd 100V condensers, 2 RF chokes, 8 B9 valve holders, 1 x 4H choke, 1 x 115V transformer, 1 boxed unit containing 4 delay lines also tag panels, trimmer condensers, suppressors, etc. on a useful chassis sized approx. 9" x 5" x 7". Only 75p (the 66 diodes would cost at least 10 times this amount). This is a snip not to be missed. Post and VAT 75p.



INFRA RED BINOCULARS

Made for military purposes during and immediately after last war to enable snipers, vehicle drivers, etc. to see in the dark. The binoculars have to be fed from a high voltage source (5kV approx.) and providing the objects are in the rays of an infra-red beam, then the binoculars will enable these objects to be seen. Each binocular eye tube contains a complete optical lens system as well as the infra-red cell, technical data which is available. The binoculars are unused, believed to be in good order. In fact they were never issued and are still in original cases, but since they were made a long time ago, they can hardly be called new. Sold without guarantee. Price £16.50 per set.



HORSTMANN 24-HOUR TIME SWITCH

With 6 position programmer. When fitted to hot water systems this could programme as follows:

Programme	Hot Water	Central Heating
0	Off	Off
1	Twice Daily	Off
2	All Day	Off
3	Twice Daily	Twice Daily
4	All Day	All Day
5	Continuously	Continuously

Suitable, of course, to programme other than central heating and hot water, for instance, programme upstairs and downstairs electric heating or heating and cooling or taped music and radio. In fact, there is no limit to the versatility of this Programmer. Mains operated. Size 3in x 3in x 2in deep. Price £5.50. Post and VAT 85p.



SWITCH TRIGGER MATS

So thin is undetectable under carpet but will switch on with slightest pressure. For burglar alarms, shop doors, etc. 24in x 18in. £1.90 plus 30p post and VAT. 13in x 10in. £1.50 plus 25p post and VAT.



Please add 30p towards packing etc. on orders under £5.

J. BULL (ELECTRICAL) LTD.

(Dept. W.W.)
103 TAMWORTH ROAD, CROYDON CRO 1XX

NEW ITEMS THIS MONTH

These and many other items were listed in our Monthly Advanced Advertising News for July. You can receive these letters and yearly catalogue by subscribing £1 yearly.

EHT transformer. American made, sealed in a steel case measuring 6 1/2" x 6" x 5 1/2" high with large porcelain stand-off insulators; it is extremely well-made, looks good enough to give 10 kv at 1 amp. Intended for American mains, its primary would have to be fed through a variac or similar. With 90v input EHT output is 6.5v. Price £15 each + carriage and VAT £3.

Sticker strip inter-phone cable. This is 4 copper wires mounted on an adhesive tape intended for telephone extensions. It can also be used for FM aerials, etc. It is invisible under thick wallpaper and of course carpet or other floor coverings. Price £5 per 100' reel + postage and VAT 60p.

2 1/2" panel meter, 0-9 amps, flush mounting, these were made for military applications, probably measurements of RF power. They work on the hot wire principle so they are suitable for AC or DC measurements. These instruments have a considerable interest and they are seldom on offer these days. Consequently they are very suitable for school labs, museums and exhibitions. Price £1.50 each + post and VAT 40p.

Instrument Mains Transformer, 6.3v at 1/2 amp and 115v at 100mA. This is an upright mounting open construction, small size (2 1/2" x 2 1/4" x 2 1/4"). Price £1.40 + postage and VAT 60p. DITD, 6.3v at 1 amp and 150v at 200mA. This is a fully shrouded upright mounting transformer, size approx. 2 1/4" x 3" x 3". Price £1.95 + post and VAT 80p.

Instrument power supply mounted on a chassis size 9" x 2 1/2" x 4 1/2". This is suitable for instruments which use valves and has AC output of 6.3v at 1 amp and a fully smoothed DC output of 150v and 200v. Price £3.00 + post and VAT 74p.

FM band amplifier. This is a 4 valve amplifier, tunable to cover FM and nearby frequencies. Tuning is by telescopic lines in fully screened compartments and the whole is mounted on a heavy metal chassis size 9" x 2 1/2" x 2 1/4" x 1". This requires a separate power supply of 150v DC at 100 mA and 6.3v at 1 amp. Price £3.75 + post and VAT £1.58.

Instrument case, grey steel finish size 9 1/2" x 3 1/4" x 2 1/4". These cases are not secondhand but come from partly assembled equipment, so the front panel has a few holes already punched through. However, these are very suitable for toggle switches, etc. There are no really big holes. 50p each + post and VAT 30p.

Instrument case, size 9" x 9" x 5 1/2", made from sheet steel grey hammer finish. No holes in these. 75p + post and VAT 49p.

PHot lamp bargain. Box of 10 x 24v. 05 amp tubular MES lamps made by Philips, the selling price of which is 25p each — our price only 50p per box.

VALVES

A1065	1.25	EF184	0.35
ARB	0.55	EFL200	0.75
ATP4	0.50	EL33	2.50
B12H	3.00	EL34	0.70
CY31	0.50	EL36	0.60
DAF96	0.50	EL41	0.80
DF96	0.55	EL81	0.60
DK96	0.55	EL82	0.55
DL92	0.40	EL84	0.30
DL96	0.40	EL85	0.60
DY82/87	0.40	EL86	0.45
DY802	0.45	EL90	0.45
EB8CC/01120	1.20	EL504	0.80
E180CC	0.70	EM31	0.60
E182CC	1.25	EM80	0.55
EASD	0.40	EM84	0.40
EA8C00	0.40	EM87	1.00
EAF42	0.75	EY51	0.45
EB91	0.30	EY81	0.45
EBF33	1.00	EY86	0.40
EBF41	0.75	EY88	0.50
EBF80	0.40	EZ40	0.70
EBF83	0.50	EZ41	0.75
EBF89	0.40	EZ80	0.30
EC52	0.35	EZ81	0.30
ECB81	0.40	GY501	0.75
ECB82	0.45	GZ34	0.70
ECB83	0.35	GZ37	1.00
ECB84	0.35	KT66	2.50
ECB85	0.40	KT88	3.20
ECB86	0.90	MH4	0.75
ECB88	0.50	ML6	0.65
ECC189	0.70	OA2	0.45
ECF80	0.40	OB2	0.45
ECF82	0.40	PAB80	0.40
ECF801	0.75	PC97	0.50
ECH42	0.80	PC900	0.50
ECH48	0.35	PCB84	0.40
ECH83	0.45	PUB21	0.70
ECH84	0.45	PCB89	0.50
ECL80	0.55	PC189	0.60
ECL82	0.35	PCF80	0.40
ECL83	0.70	PCF82	0.40
ECL86	0.50	PCF84	0.60
EF36	0.65	PCF86	0.40
EF37A	1.20	PCF200	0.75
EF40	0.75	PCF201	0.75
EF41	0.65	PCF801	0.55
EF80	0.30	PCF802	0.50
EF83	1.25	PCF805	0.90
EF85	0.35	PCF806	0.75
EF86	0.35	PCF808	0.90
EF89	0.30	PCH200	0.80
EF91	0.45	PCL81	0.55
EF92	0.50	PCL82	0.40
EF95	0.40	PCL83	0.45
EF183	0.35	PCL84	0.65

PCL86	0.50	Z801U	2.70	6AK5	0.40	RCL6	0.75	6X5GT	0.50
PCL805	0.60	Z900T	1.20	6AK6	0.40	6D6	0.55	6V6G	0.90
PL200	0.70	1A3	0.55	6AL5	0.30	6EAB	0.85	6Z4	0.65
PL36	0.60	1L4	0.25	6AL5W	0.55	6F7	1.10	6Z0L2	0.90
PL81	0.50	3A4	0.60	6AM6	0.45	6F8G	0.75	7B7	0.80
PL82	0.45	1R5	0.40	6AN8	0.45	6F23	0.90	7Y4	0.80
PL83	0.45	1S4	0.30	6AQ5	0.45	6F32	0.75	9D6	0.40
PL84	0.45	1T4	0.30	6AQ5W	0.70	6F33	3.50	9D2	0.60
						6H6	0.35	12AG	0.55
						6J4WA	1.25	12A6	0.45
						6J5	0.65	12AT7	0.40
						6J5GT	0.50	2A7	0.35
						6J6	0.30	12AV6	0.50
						6J7	0.60		
						6J7G	0.40		
						6K6GT	0.80		
						6K7	0.55		
						6K7G	0.30		
						6K8GT	0.40		
						6K25	1.00		
						6L6	1.90		
						6L9G	0.60		
						6AXAGT	0.75		
						6AX5GT	1.00		
						6B7	0.70		
						6BA6	0.35		
						6BE6	0.40		
						6BGGG	0.90		
						6B36	0.65		
						6J5GT	0.40		
						6S4	0.70		
						6S4GT	0.50		
						6S7	0.50		
						6S7GT	0.50		
						6T2E1	3.50		
						12K5	1.10		
						12K7GT	0.50		
						12X8GT	0.70		
						12Q7G	0.70		
						12SA7GT	0.70		
						12S7G	0.55		
						12S7J	0.55		
						12V4	0.40		
						14S7	1.00		
						19A05	6.55		
						19G3	8.00		
						19G6	6.60		
						19H5	14.00		
						20P3	0.60		
						20R4	1.10		
						25L6GT	0.70		
						30P15	1.00		
						30C18	0.90		
						30F5	1.00		
						30FL1	1.00		
						30FL12	1.10		
						30FL14	0.90		
						30L15	0.95		
						30L17	0.95		
						30P12	1.00		
						30P19	1.00		
						30P1	0.95		
						30P13	1.10		

VALVES AND TRANSISTORS

Telephone enquiries for valves, transistors, etc. retail 749 3934, trade and export 743 0899

5V4G	0.55	6BQ7A	0.60	6SJ7GT	0.35	30P14	1.10
5Y3GT	0.85	6BR7	1.20	6SK7	0.55	35L6GT	0.75
5Z3	0.80	6BW6	1.00	6SL7GT	0.50	35W4	0.50
5Z4	0.80	6BW7	1.00	6SN7GT	0.50	35Z4GT	0.70
5Z4TG	0.85	6C4	0.40	6S9GT	0.55	50C5	0.60
6AB7	0.90	6C6	0.50	6V6GT	0.50	50D6B	1.10
6AC7	0.60	6C8	0.50	6X4	0.40	75C1	0.75
6AH6	0.70	6C6H	1.45	6X5G	0.40		

TRANSISTORS

Please write or phone for current price of any of the transistors, diodes shown below.

AF178	BF167	GET115	OC36	Sx754	2N2062
AF186	BF185	GET116	OC42	ZR11	2N2147
AF212	BFY51	GEX66	OC44	ZR21	2N2147
ASV26	BFY52	NKT222	OC45	1N238	2N2989
ASV27	BFY90	OA5	OC70	1N25	2N3053
AD161	BSY27	OA47	OC73	1N32A	2N3054
BC108	BSY38	OA70	OC78	1N38A	2N3055
BC118	BSY95A	OA71	OC7B	1N43	2N3200
BC119	BY716	OA73	OC81	1N70	2N3391
AF114	BC136	CRS1 10	OA79	OC82	2N5295
AF115	BC137	CRS1 20	OA91	OC82D	3N128
AF116	BC148A	CRS1 30	OA200	OC83	3N154

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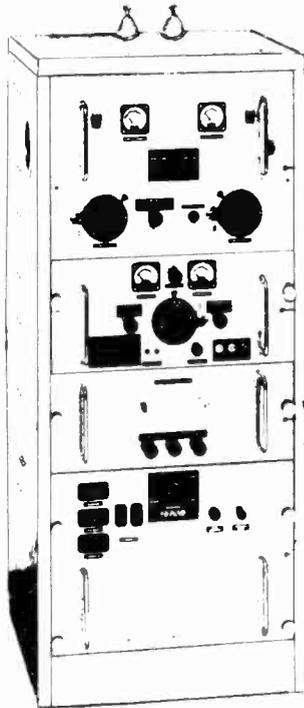
0B2 0.50	6AX4 0.88	6K7G 0.35	12A16 0.53	30FL14 0.82	AC2P/END 1.17	EA8F801 0.80	EF36 0.95	KTW62 1.78	PL33 0.47	UF89 0.47	X65 1.46	AD161 0.59	BY127 0.23	0A200 0.12
0B4 0.55	6B8G 0.35	6K8G 0.33	12A17 0.39	30L1 0.35	AC6/PEN 0.60	EF34 0.40	EF183 0.40	PL84 0.50	PL84 0.50	UL41 0.75	X68 0.85	AD162 0.59	BY223 1.29	0A202 0.13
1A3 0.65	6BA6 0.41	6L1 2.34	12A16 0.59	30L15 0.82	AC/PEN/7 1.17	EB281 0.45	EF194 0.40	PL302 0.88	UL84 0.49	Y84 0.50	X61 0.85	AD163 1.16	BY210 0.33	0A210 0.62
1A3T 0.59	6BC8 0.90	6L6G 0.68	12AX7 0.39	30L17 0.78	AC/TH/1 1.00	EB282 0.45	EF804 1.75	MHL06 1.00	UL84 0.49	Y84 0.50	X61 0.85	AD164 1.16	BY211 0.33	0A211 0.86
1B3GT 0.59	6B6E 0.41	6L7(M) 0.59	12AY7 0.94	30P4MR 1.05	AL60 1.17	EB283 0.45	EH9 0.53	PL505 1.85	UL84 0.49	Y84 0.50	X61 0.85	AD165 1.16	BY212 0.33	0A212 0.86
1C2 1.15	6B86 0.75	6L18 0.64	12B6E 0.50	30P12 1.05	AL60 1.17	EB284 0.45	EK90 0.53	PL506 1.10	UL84 0.49	Y84 0.50	X61 0.85	AD166 1.16	BY213 0.33	0A213 0.86
1HG 1.17	6B86 0.64	6L19 2.34	12BH7 0.59	30P19/ 1.05	AL60 1.17	EB285 0.45	EL32 0.60	PL508 1.10	UL84 0.49	Y84 0.50	X61 0.85	AD167 1.16	BY214 0.33	0A214 0.86
1I56T 0.90	6BK7A 0.85	6L1D2 0.45	12BY7 0.85	30P4 0.88	AL60 1.17	EB286 0.45	EL33 3.00	PL509 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD168 1.16	BY215 0.33	0A215 0.86
1L4 0.30	6BQ5 0.36	6L1D20 0.88	12E1 3.51	30P16 0.43	AL60 1.17	EB287 0.45	EL34 1.00	PL509 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD169 1.16	BY216 0.33	0A216 0.86
1LD5 0.70	6BQ7A 0.64	6N7GT 0.70	12J5GT 0.39	30P18 0.50	AL60 1.17	EB288 0.45	EL35 3.00	PL510 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD170 1.16	BY217 0.33	0A217 0.86
1LN5 0.70	6BR7 1.20	6PL12 0.45	12J7GT 0.70	30P11 1.00	AL60 1.17	EB289 0.45	EL36 3.00	PL511 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD171 1.16	BY218 0.33	0A218 0.86
1NSGT 0.76	6BR8 1.25	6P15 0.36	12K5 1.17	30P12 0.45	AL60 1.17	EB290 0.45	EL37 3.00	PL512 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD172 1.16	BY219 0.33	0A219 0.86
1R5 0.50	6BS7 1.64	6Q07 0.60	12K7GT 0.59	30P13 1.20	AL60 1.17	EB291 0.45	EL38 3.00	PL513 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD173 1.16	BY220 0.33	0A220 0.86
1S4 0.39	6BW6 1.00	6Q2GT 0.60	12K8 0.85	30P14 1.29	AL60 1.17	EB292 0.45	EL39 3.00	PL514 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD174 1.16	BY221 0.33	0A221 0.86
1S5 0.40	6BW7 0.65	6Q2(M) 0.64	12Q7GT 0.53	30P15 0.84	AL60 1.17	EB293 0.45	EL40 3.00	PL515 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD175 1.16	BY222 0.33	0A222 0.86
1T4 0.35	6BK8 0.30	6R7G 0.70	12SA7GT 0.70	35A3 0.76	AL60 1.17	EB294 0.45	EL41 0.60	PL516 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD176 1.16	BY223 0.33	0A223 0.86
1U4 0.70	6BY7 0.40	6R7(M) 0.88	12S7GT 0.70	35A3 0.76	AL60 1.17	EB295 0.45	EL42 0.60	PL517 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD177 1.16	BY224 0.33	0A224 0.86
1U5 0.88	6BZ6 0.57	6SA7 0.55	12S7GT 0.70	35A3 0.76	AL60 1.17	EB296 0.45	EL43 0.60	PL518 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD178 1.16	BY225 0.33	0A225 0.86
2D21 0.60	6C4 0.47	6SCTGT 0.50	12S7GT 0.70	35A3 0.76	AL60 1.17	EB297 0.45	EL44 0.60	PL519 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD179 1.16	BY226 0.33	0A226 0.86
2GK5 0.75	6C5G 0.50	6S7G 0.52	12S7GT 0.70	35A3 0.76	AL60 1.17	EB298 0.45	EL45 0.60	PL520 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD180 1.16	BY227 0.33	0A227 0.86
2X2 0.70	6C6 0.47	6S7 0.55	12S7GT 0.70	35A3 0.76	AL60 1.17	EB299 0.45	EL46 0.60	PL521 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD181 1.16	BY228 0.33	0A228 0.86
3A4 0.70	6C7 2.00	6S7 0.55	12S7GT 0.70	35A3 0.76	AL60 1.17	EB300 0.45	EL47 0.60	PL522 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD182 1.16	BY229 0.33	0A229 0.86
3B7 0.53	6C8 0.95	6S7GT 0.52	12SN7GT 0.70	35A3 0.76	AL60 1.17	EB301 0.45	EL48 0.60	PL523 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD183 1.16	BY230 0.33	0A230 0.86
3B28 4.50	6C8A 0.47	6S7GT 0.52	12SN7GT 0.70	35A3 0.76	AL60 1.17	EB302 0.45	EL49 0.60	PL524 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD184 1.16	BY231 0.33	0A231 0.86
3D6 0.47	6C12 0.40	6U4GT 0.82	12SQGT 0.70	35A3 0.76	AL60 1.17	EB303 0.45	EL50 0.60	PL525 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD185 1.16	BY232 0.33	0A232 0.86
3Q4 0.85	6C17 2.34	6U7G 0.55	12T7GT 0.70	35A3 0.76	AL60 1.17	EB304 0.45	EL51 0.60	PL526 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD186 1.16	BY233 0.33	0A233 0.86
3Q5GT 0.70	6C18 1.60	6V6G 0.30	12SR7 0.75	35A3 0.76	AL60 1.17	EB305 0.45	EL52 0.60	PL527 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD187 1.16	BY234 0.33	0A234 0.86
3S4 0.47	6C6A 0.88	6V6GT 0.53	14H7 0.64	35A3 0.76	AL60 1.17	EB306 0.45	EL53 0.60	PL528 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD188 1.16	BY235 0.33	0A235 0.86
3V4 0.82	6CL6 0.76	6X4 0.47	14S7 1.10	35A3 0.76	AL60 1.17	EB307 0.45	EL54 0.60	PL529 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD189 1.16	BY236 0.33	0A236 0.86
4CB6 0.75	6CL8A 0.94	6X5GT 0.53	18 1.17	35A3 0.76	AL60 1.17	EB308 0.45	EL55 0.60	PL530 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD190 1.16	BY237 0.33	0A237 0.86
5CG8 0.75	6CM7 0.88	6Y6G 0.94	19A9G 0.65	35A3 0.76	AL60 1.17	EB309 0.45	EL56 0.60	PL531 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD191 1.16	BY238 0.33	0A238 0.86
5R4G 0.94	6C5 0.55	6Y7 1.17	19B6G 0.65	35A3 0.76	AL60 1.17	EB310 0.45	EL57 0.60	PL532 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD192 1.16	BY239 0.33	0A239 0.86
5T4 0.47	6C4W 1.17	7A7 1.00	1.17	35A3 0.76	AL60 1.17	EB311 0.45	EL58 0.60	PL533 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD193 1.16	BY240 0.33	0A240 0.86
5U4G 0.50	6D3 0.75	7B6 1.88	19G6 7.00	35A3 0.76	AL60 1.17	EB312 0.45	EL59 0.60	PL534 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD194 1.16	BY241 0.33	0A241 0.86
5V4G 0.59	6D7 0.88	7B7 0.82	19H1 4.00	35A3 0.76	AL60 1.17	EB313 0.45	EL60 0.60	PL535 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD195 1.16	BY242 0.33	0A242 0.86
5Y3GT 0.55	6D8A 0.88	7B8 1.76	20D1 0.80	35A3 0.76	AL60 1.17	EB314 0.45	EL61 0.60	PL536 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD196 1.16	BY243 0.33	0A243 0.86
5Z3 0.88	6E6W 0.88	7H7 0.88	20D4 2.34	35A3 0.76	AL60 1.17	EB315 0.45	EL62 0.60	PL537 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD197 1.16	BY244 0.33	0A244 0.86
5Z4G 0.55	6E5 1.17	7V7 1.76	20F2 0.88	35A3 0.76	AL60 1.17	EB316 0.45	EL63 0.60	PL538 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD198 1.16	BY245 0.33	0A245 0.86
5Z4GT 0.55	6F1 0.80	7Y4 0.88	20L1 1.29	35A3 0.76	AL60 1.17	EB317 0.45	EL64 0.60	PL539 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD199 1.16	BY246 0.33	0A246 0.86
6/30L2 0.80	6F6G 0.60	7Z4 0.94	20P1 0.64	35A3 0.76	AL60 1.17	EB318 0.45	EL65 0.60	PL540 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD200 1.16	BY247 0.33	0A247 0.86
6A8G 1.46	6F12 0.50	9B5W 0.88	20P3 0.94	35A3 0.76	AL60 1.17	EB319 0.45	EL66 0.60	PL541 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD201 1.16	BY248 0.33	0A248 0.86
6A7 0.80	6F13 0.80	9D7 0.76	20P4 1.17	35A3 0.76	AL60 1.17	EB320 0.45	EL67 0.60	PL542 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD202 1.16	BY249 0.33	0A249 0.86
6AG5 0.35	6F14 0.88	10C2 0.76	20P5 1.50	35A3 0.76	AL60 1.17	EB321 0.45	EL68 0.60	PL543 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD203 1.16	BY250 0.33	0A250 0.86
6AH6 0.80	6F15 0.76	10D1 0.82	25A6G 0.70	35A3 0.76	AL60 1.17	EB322 0.45	EL69 0.60	PL544 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD204 1.16	BY251 0.33	0A251 0.86
6AJ5 0.76	6F18 0.84	10DE7 0.88	25L6G 0.70	35A3 0.76	AL60 1.17	EB323 0.45	EL70 0.60	PL545 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD205 1.16	BY252 0.33	0A252 0.86
6AJ8 0.40	6F23 0.82	10F1 0.88	25V5 0.80	35A3 0.76	AL60 1.17	EB324 0.45	EL71 0.60	PL546 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD206 1.16	BY253 0.33	0A253 0.86
6AK5 0.47	6F24 1.00	10F3 1.17	25V5G 0.80	35A3 0.76	AL60 1.17	EB325 0.45	EL72 0.60	PL547 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD207 1.16	BY254 0.33	0A254 0.86
6AK6 0.70	6F25 1.17	10F9 0.76	25Z4G 0.50	35A3 0.76	AL60 1.17	EB326 0.45	EL73 0.60	PL548 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD208 1.16	BY255 0.33	0A255 0.86
6AK8 0.45	6F26 0.40	10F18 0.64	25Z5 0.75	35A3 0.76	AL60 1.17	EB327 0.45	EL74 0.60	PL549 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD209 1.16	BY256 0.33	0A256 0.86
6AL5 0.23	6F28 0.78	10L14 0.53	25Z6G 0.80	35A3 0.76	AL60 1.17	EB328 0.45	EL75 0.60	PL550 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD210 1.16	BY257 0.33	0A257 0.86
6AM8A 0.70	6F32 0.70	10LD11 0.82	28D7 2.00	35A3 0.76	AL60 1.17	EB329 0.45	EL76 0.60	PL551 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD211 1.16	BY258 0.33	0A258 0.86
6AN8 0.82	6G6G 0.60	10LP12 0.45	30A5 0.76	35A3 0.76	AL60 1.17	EB330 0.45	EL77 0.60	PL552 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD212 1.16	BY259 0.33	0A259 0.86
6AQ5 0.53	6G8A 0.88	10P13 0.88	30C1 0.47	35A3 0.76	AL60 1.17	EB331 0.45	EL78 0.60	PL553 1.65	UL84 0.49	Y84 0.50	X61 0.85	AD213 1.16	BY260 0.33	0A260 0.86
6AQ8 0.47	6GK5 0.78	10P14 2.34	30C1S 0.82	35A3 0.76	AL60 1.17	EB332 0.45	EL79 0.60	PL554 1.65	UL84 0.4					

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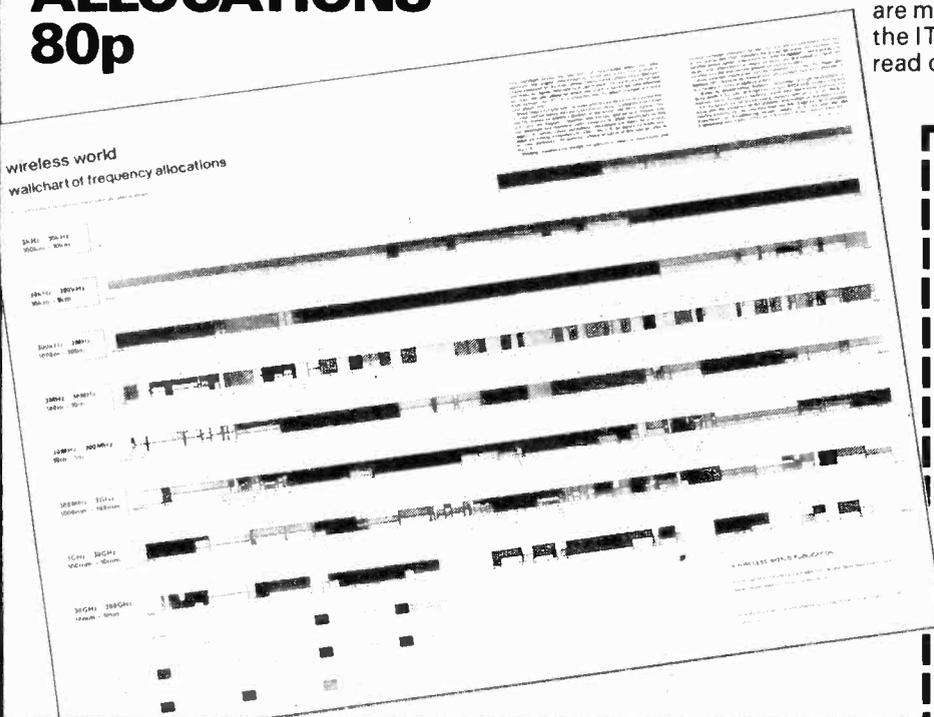
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1 6	12	6	5.52	53
7	16	8	7.28	60
1 5	20	10	10.39	73
1 7	30	15	13.59	83
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79	1.0	2.40	38
3	2.0	3.49	38
20	3.0	4.53	45
21	4.0	5.13	53
51	5.0	6.41	53
117	6.0	7.16	60
88	8.0	9.87	67
89	10.0	9.90	73

50 VOLT RANGE

SECONDARY TAPS 0-19-25-33-40-50

Ref. No.	Amps	£	P&P p
102	0.5	2.58	30
103	1.0	3.38	38
104	2.0	4.68	45
105	3.0	5.81	53
106	4.0	7.60	67
107	6.0	12.10	67
118	8.0	12.98	85
119	10.0	16.99	BRS

60 VOLT RANGE

SECONDARY TAPS 0-24-30-40-48-60

Ref. No.	Amps	£	P&P p
124	0.5	2.33	38
126	1.0	3.41	38
127	2.0	5.08	45
125	3.0	7.52	60
123	4.0	8.75	67
40	5.0	9.75	73
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131	8.0	15.00	BRS
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64	75	0-115-210-240	2.90	38
4	150	0-115-210-220-240	4.12	45
66	300	"	5.82	53
67	500	"	8.82	67
84	1000	"	13.68	91
93	1500	"	18.11	BRS
95	2000	"	24.20	BRS
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Ref.	MA	Volts	£	P&P p
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212	1A1A	0-6-0-6	1.84	30
13	100	9-0-9	1.41	13
235	330 330	0-9-0-9	1.56	19
207	500 500	0-8-9-0-8-9	1.92	30
208	1A A	0-8-9-0-8-9	3.30	38
236	20C 200	0-15-0-15	1.43	19
214	30C 300	0-20-0-20	1.93	30
221	70C (DC)	20-12-0-12-20	2.17	38
206	1A A	0-15-20-0-15-20	3.46	38
203	50C 500	0-15-27-0-15-27	3.00	38
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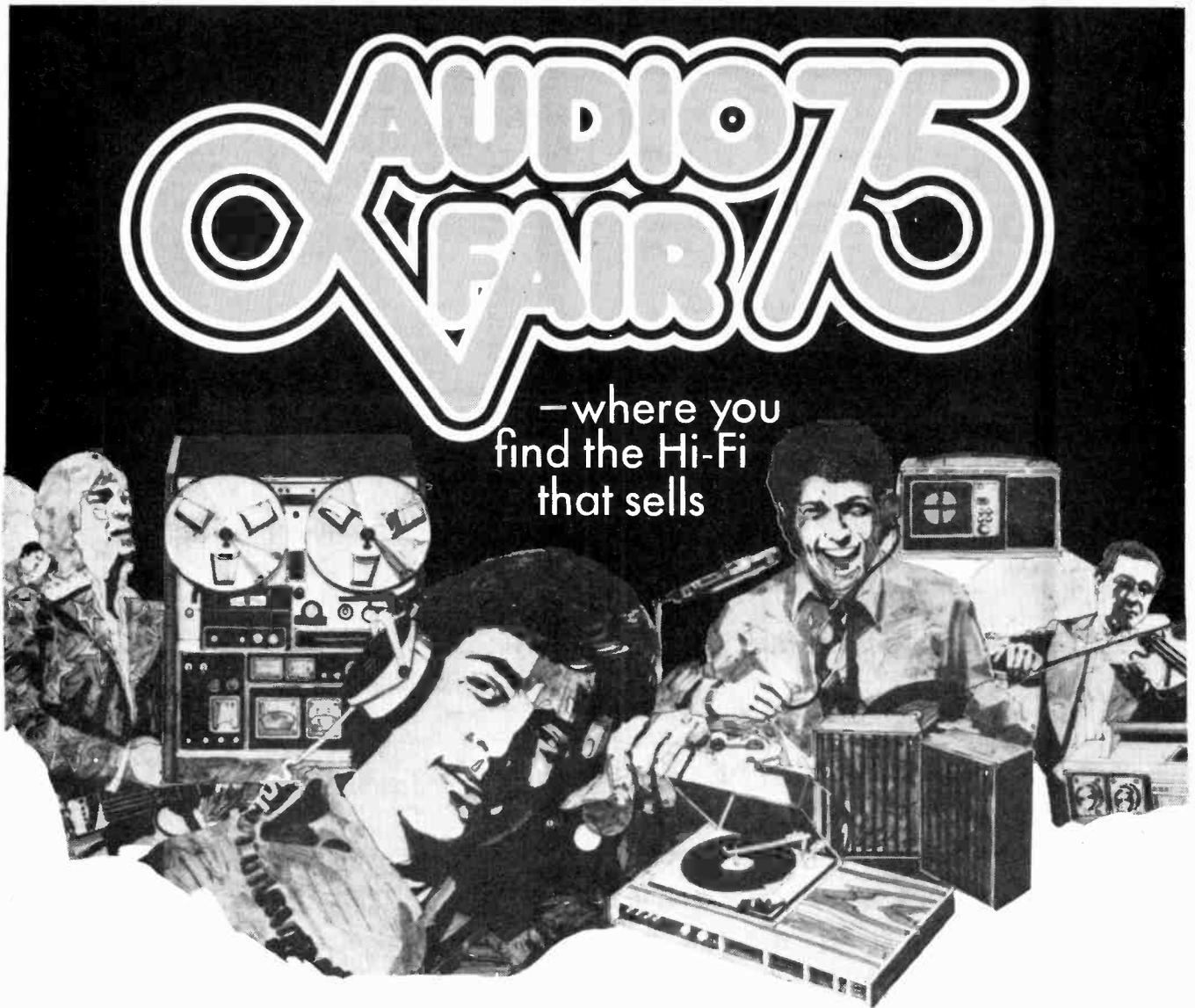
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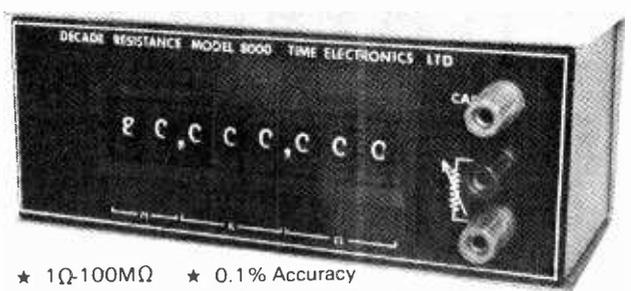


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TRANSISTORS		Type	Price (p)	Type	Price (p)	Type	Price (p)	Type	Price (p)	Type	Price (p)
AC107	0.35	BC119	0.29	BD123	0.85	BF273	0.16	CE106F	0.43	TX310	0.10
AC111	0.20	BC125	0.22	BD131	0.98	BF336	0.35	CE111E	0.56	TX313	0.12
AC126	0.25	BC126	0.20	BD130Y	1.42	BF458	0.60	CRS1/40	0.75	TX500	0.17
AC127	0.25	BC132	0.15	BD131	0.45	BF459	0.63	CRS3/40	0.95	TX502	0.17
AC128	0.25	BC134	0.20	BD135	0.40	BF597	0.70	D40N1	0.45	TX504	0.42
AC141	0.26	BC135	0.19	BD136	0.46	BF597	0.15	E1222	0.55	TX602	0.24
AC141K	0.27	BC136	0.20	BD137	0.48	BF639	0.24	E5024	0.20	2N526	0.86
AC142K	0.19	BC137	0.20	BD138	0.50	BF681	0.30	ME6001	0.16	2N526	0.18
AC153K	0.28	BC138	0.20	BD139	0.55	BF681	0.30	ME6002	0.17	2N526	0.15
AC154	0.20	BC142	0.30	BD140	0.38	BF743	0.55	MJE8001	1.18	2N706	0.12
AC176	0.25	BC143	0.35	BD144	2.19	BFW10	0.55	MJE340	0.68	2N706A	0.15
AC178	0.27	BC147B	0.17	BD147B	0.17	BF127	0.20	MJE341	0.72	2N708	0.36
AC179	0.27	BC148	0.12	BD163	0.67	BFW18A	1.70	MJE370	0.85	2N744	0.30
AC187	0.25	BC149	0.14	BD183	0.56	BFW30	1.38	MJE370	0.85	2N914	0.19
AC187K	0.26	BC149B	0.15	BD234	0.75	BFW59	0.19	MJE2955	1.20	2N918	0.42
AC188	0.25	BC152	0.25	BD519	0.76	BFW60	0.20	MJE3000	1.85	2N930	0.35
AC188K	0.26	BC153	0.20	BD520	0.76	BFW90	0.28	MJE3055	0.74	2N1304	0.21
AC 193K	0.30	BC154	0.20	BOX18	1.45	BFX16	2.55	MM721	0.70	2N1305	0.21
AC194K	0.32	BC157	0.15	BOX32	2.55	BFX29	0.30	MPF102	0.40	2N1306	0.31
ACV28	0.25	BC158	0.13	BDY16A	0.38	BFX30	0.35	MPSA05	0.47	2N1307	0.22
ACV39	0.68	BC159	0.15	BDY18	1.78	BFX84	0.25	MPSA55	0.50	2N1308	0.26
AD140	0.50	BC161	0.48	BDY20	0.99	BFX85	0.28	MPSU05	0.66	2N1513	0.34
AD 142	0.52	BC167B	0.15	BF115	0.20	BFY06	0.26	MPSU06	0.76	2N1513	0.45
AD143	0.51	BC168B	0.13	BF120	0.25	BFY07	0.28	MPSU06	1.28	2N1890	0.45
AD149	0.48	BC169C	0.13	BF120	0.55	BFX88	0.24	MPSU56	0.76	2N1893	0.48
AD161	0.48	BC170	0.15	BF121	0.25	BFY18	0.53	OC26	0.38	2N2102	0.51
AD162	0.48	BC171A	0.15	BF123	0.28	BFY40	0.40	OC28	0.85	2N2217	0.36
AF114	0.25	BC172	0.14	BF125	0.25	BFY41	0.43	OC35	0.59	2N2218	0.60
AF115	0.25	BC173	0.20	BF127	0.20	BFY42	0.43	OC36	0.59	2N2219	0.50
AF116	0.25	BC176	0.12	BF158	0.25	BFY51	0.23	OC42	0.55	2N2218A	0.41
AF117	0.20	BC177	0.20	BF159	0.27	BFY52	0.23	OC44	0.25	2N2222A	0.50
AF118	0.20	BC178	0.22	BF160	0.22	BFY57	0.32	OC45	0.32	2N2269A	0.42
AF121	0.32	BC178B	0.22	BF161	0.45	BFY64	0.42	OC70	0.32	2N2401	0.60
AF124	0.29	BC179	0.20	BF162	0.45	BFY72	0.31	OC71	0.32	2N2484	0.41
AF125	0.25	BC179B	0.13	BF163	0.45	BFY90	0.70	OC72	0.32	2N2484	0.18
AF126	0.25	BC182L	0.11	BF167	0.25	BLY15A	0.79	OC73	0.51	2N2646	0.53
AF127	0.25	BC183	0.11	BF173	0.25	BPX25	1.90	OC75	0.25	2N2712	0.12
AF139	0.35	BC183K	0.12	BF177	0.30	BPX29	1.70	OC81	0.57	2N2904	0.22
AF147	0.35	BC183L	0.11	BF178	0.33	BPX52	1.90	OC81D	0.53	2N2904A	0.26
AF149	0.45	BC184L	0.13	BF179	0.33	BRCA443	0.68	OC139	0.76	2N5496	1.05
AF178	0.55	BC187	0.17	BF181	0.33	BRCA443	0.68	OC140	0.76	2N5496	1.05
AF179	0.60	BC187	0.17	BF181	0.33	BRCA443	0.68	OC170	0.25	2N2905A	0.28
AF180	0.55	BC208	0.12	BF182	0.44	BR101	0.47	OC171	0.30	2N2926G	0.13
AF181	0.50	BC212L	0.12	BF183	0.44	BSW64	0.38	OC171	0.30	2N2926G	0.12
AF186	0.40	BC213L	0.12	BF184	0.26	BSX19	0.13	ON188	1.19	2N3019	0.75
AF239	0.40	BC214L	0.15	BF185	0.26	BSX20	0.15	ON236A	0.85	2N3019	0.21
AF279	0.64	BC238	0.12	BF194	0.15	BSX76	0.15	ORP12	1.25	2N3054	0.55
AL100	1.10	BC261A	0.28	BF195	0.15	BSX82	0.52	R2008B	2.05	2N3055	0.50
AL102	1.10	BC262A	0.18	BF196	0.15	BSY19	0.52	R2010B	2.95	2N3133	0.54
AL103	1.10	BC263B	0.25	BF197	0.17	BSY41	0.22	TAG3/400		2N3134	0.60
AL113	0.95	BC267	0.16	BF198	0.20	BSY52	0.45		1.54	2N3232	1.32
AU103	2.10	BC268C	0.14	BF199	0.25	BSY54	0.50	TIC44	0.29	2N3235	1.10
AU110	1.90	BC270	0.33	BF200	0.35	BSY56	0.80	TIC46	0.44	2N3254	0.28
AU113	2.40	BC300	0.60	BF218	0.35	BSY65	0.15	TIC47	0.58	2N3323	0.48
BC107	0.12	BC301	0.35	BF222	1.08	BSY78	0.40	TIC29A	0.49	2N3391A	0.23
BC107B	0.40	BC303	0.60	BF224	1.15	BSY91	0.28	TIP30A	0.58	2N3702	0.13
BC108	0.12	BC307B	0.12	BF240	0.20	BT106	1.24	TIP31A	0.65	2N3703	0.15
BC108A	0.12	BC308A	0.16	BF241	0.22	BT116	1.20	TIP32A	0.87	2N3704	0.18
BC108B	0.13	BC309	0.15	BF244	0.18	BU105/021	95	TIP33A	0.99	2N3705	0.11
BC108C	0.14	BC323	0.68	BF254	0.45	BU108	3.25	TIP34A	1.73	2N3706	0.10
BC109	0.13	BC377	0.22	BF255	0.45	BU126	2.99	TIP41A	0.80	2N3707	0.13
BC109C	0.14	BC441	1.10	BF256	0.45	BU204	1.98	TIP42A	0.91	2N3715	0.20
BC113	0.13	BC461	1.58	BF257	0.49	BU209	3.08	TIS43	0.30	2N3715	0.20
BC114	0.20	BC476	1.15	BF258	0.49	BU207	3.08	TIS43	0.30	2N3715	0.20
BC115	0.20	BCY71	0.22	BF259	0.93	BU208	3.15	ZTX109	0.15	2N3717	1.70
BC116	0.20	BCY78	4.65	BF262	0.70	BU209	2.55	ZTX300	0.16	2N3772	1.90
BC117	0.20	BCY88	2.42	BF263	0.70	BU277	2.55	ZTX304	0.22	2N3773	2.90

DIODES

Type	Price (p)	Type	Price (p)
2N3780	4.15	2N3780	0.20
AA113	0.15	2N3819	0.35
AA119	0.09	2N3820	0.49
AA129	0.20	2N3823	1.45
AA143	0.10	2N3866	1.70
AAZ13	0.30	2N3877	0.25
AAZ17	0.12	2N3904	0.18
BA100	0.15	2N3905	0.18
BA102	0.25	2N3906	0.15
BA110U	0.30	2N4032	0.43
BA115	0.12	2N4036	0.52
BA141	0.17	2N4046	0.35
BA145	0.17	2N4058	0.17
BA148	0.15	2N4123	0.13
BA154	0.13	2N4124	0.15
BA155	0.16	2N4126	0.20
BA156	0.15	2N4236	1.80
BA157	0.25	2N4248	0.12
BA163	0.08	2N4284	0.19
BA166	0.07	2N4288	0.13
BA167	0.11	2N4289	0.20
BA168	0.11	2N4290	0.18
BA169	0.11	2N4291	0.18
BA170	0.11	2N4292	0.20
BA171	0.11	2N4293	0.24
BA172	0.11	2N4294	1.30
BA173	0.11	2N4295	1.05
BA174	0.11	2N4296	0.32
BA175	0.11	2N4297	0.35
BA176	0.11	2N4298	0.45
BA177	0.11	2N4299	0.32
BA178	0.11	2N4300	0.35
BA179	0.11	2N4301	0.35
BA180	0.11	2N4302	0.35
BA181	0.11	2N4303	0.35
BA182	0.11	2N4304	0.35
BA183	0.11	2N4305	0.35
BA184	0.11	2N4306	0.35
BA185	0.11	2N4307	0.35
BA186	0.11	2N4308	0.35
BA187	0.11	2N4309	0.35
BA188	0.11	2N4310	0.35
BA189	0.11	2N4311	0.35
BA190	0.11	2N4312	0.35
BA191	0.11	2N4313	0.35
BA192	0.11	2N4314	0.35
BA193	0.11	2N4315	0.35
BA194	0.11	2N4316	0.35
BA195	0.11	2N4317	0.35
BA196	0.11	2N4318	0.35
BA197	0.11	2N4319	0.35
BA198	0.11	2N4320	0.35
BA199	0.11	2N4321	0.35
BA200	0.11	2N4322	0.35

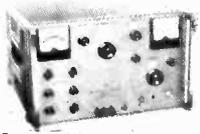
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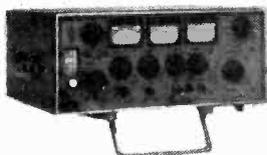
- Unit Oscillator 1218-A P.O.A.
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MARCONI INTS.

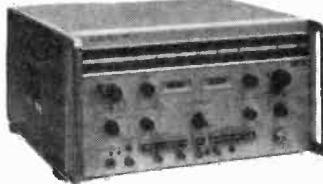
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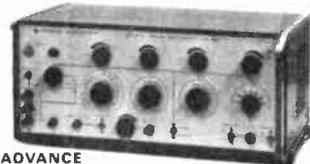


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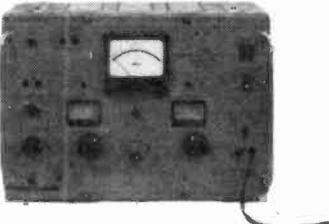
MARCONI

- Double Pulse Generator TF 14D0/S. Min type CT 434 c/w TM 6600 Sec. Pulse gen Plug In. 10Hz-100KHz 100nsec-100 μ sec. Negative pulses up to 200V E.M.F. +VE pulses up to 60V E.M.F. & Simultaneous +VE & -VE pulses up to 20V E.M.F. P.O.A.

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S.T.C.

- Octave Filter 74143A. 37.5-12,800Hz For analysing noise and interference on comms. systems, particularly useful with 74142 osophometer P.O.A.
- Selective level Measuring Set 74184B 60-1364KHz P.O.A.
- Measuring Set 74831A P.O.A.

A.T.&E.

- Telegraph Distortion Measuring Set. Various types 5BV, 5BV3, 6BV, 6A P.O.A.

MURPHY RADIO

- Receiver VHF field strength RX 506 & interference measuring set c/w Power Unit P.O.A.

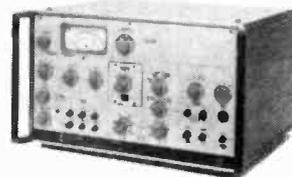
OSCILLOSCOPE TEST EQUIPMENT



TEKTRONIX

- Pulse Generator Type III. 0.5 μ sec risetime. Amplitude \pm 5V. 30 to 250 nsec. difference between trigger and output pulses **£60**
- Time Mark Generator 180A **£85**
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- Pre-Trigger Pulse Generator III **£60**

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MARCONI

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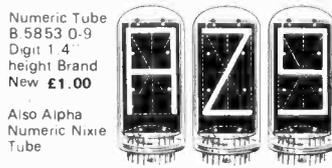
GRESHAM LION

- Waveform Generator 625 lines. Sine squared. Pulse & Bar P.O.A.
- Waveform Generator 625 lines staircase P.O.A.

WANDEL & GOLTERMAN

- Measuring Set VZM 1 **£495**
- Generator & Receiver. VZM 83 **£275**

COMPONENTS

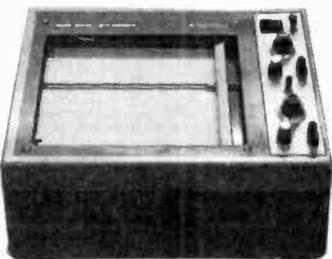


- Numeric Tube B 5853 0-9 Digit 1.4" height Brand New **£1.00**

- Also Alpha Numeric Nixie Tube

- B.7971. Displays alphabet & 0-9 numerals **99p**

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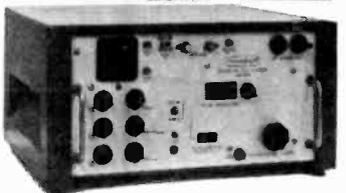
- ADVANCE X-Y Recorder HR 92/1 **£150**

- BELL & HOWELL U.V. Recorder 5/127. without Galvo's **£225**
- BRYANS X-Y Recorder 20021 **£195**
- ELECTRONIC ASSOCIATES Vari-Plotter **£145**
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- EVERSHED & VIGNOLES Circuit Tester Ohmmeter 0-3 ohms 0-39 ohms **£18**
- Circuit Tester Ohmmeter 0-1000 ohms 100-200 kohms **£18**
- Megger Series +II 250V **£12**
- Megger Series III Mk. 3 250V **£20**
- Megger 250V **£20**
- Megger 500V **£37**
- Battery Megger 500V **£37**
- PYE Insulation Tester 10B 500V **£25**

MISCELLANEOUS



- Transfer Oscillator Type 75B0H by Beckman. DC-15GHz with counter. 7.5MHz-15GHz without counter. Sensitivity 100mV (R.M.S.) **£350**

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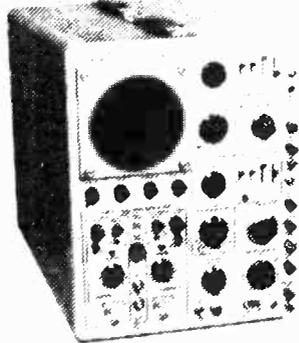
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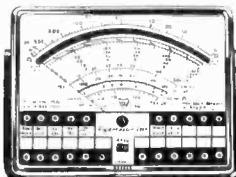
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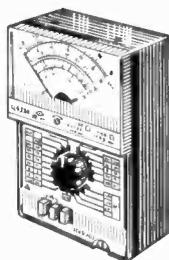
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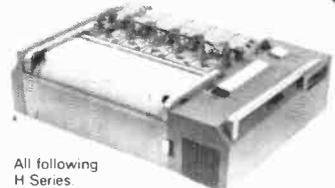
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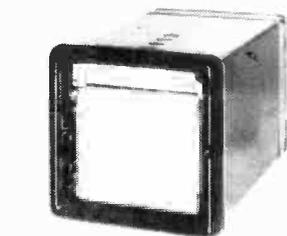
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- check tape switch for encoded monitoring in three-head machines

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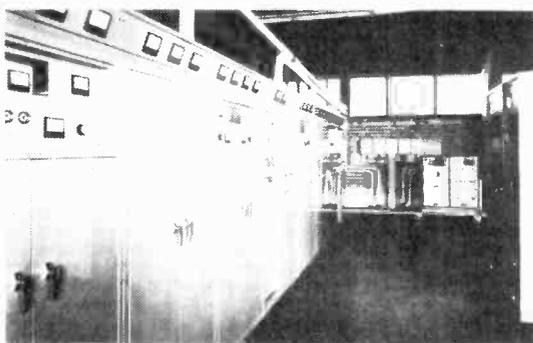


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(4873)



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We are looking for a first class engineer to develop and expand our power electronics range of instruments. The successful applicant will lead an exciting activity in power control using thyristors, triacs, power transistors and associated low signal control circuits. He will be responsible for the development of new and existing products and their application in the industrial field.

The man we seek should have several years experience of design, he will be an ideas man with an HNC, degree or its equivalent who likes challenge and responsibility. We can promise him job satisfaction and good remuneration as well as a place in a well integrated expanding team. If you think that this will be of interest to you then please phone for an informal discussion with the Development Manager, Mr. W. G. Ashman on Stevenage 4422 or write to the Personnel Manager at



Pye Ether Ltd
Caxton Way, Stevenage, Herts, SG1 2DG
 Tel: Stevenage (0438) 4422

4878

TEST AND CALIBRATION ENGINEERS

We have vacancies for

TEST ENGINEERS to fault-find and test a wide variety of electronic control and nucleonic equipment.

CALIBRATION ENGINEERS with experience in the maintenance, repair and calibration of our high-grade electronic test and laboratory equipment.

Academic qualifications, whilst desirable, are less important than sound experience. Minimum age 25 years. These positions would be ideal for ex-service men.

Good rates of pay, 4 weeks holiday, pension and sick pay schemes.

Ring Sylvia Borra 01-692 1271 Ext 393

or write to her at

The Personnel Department

(4898)

GEC-ELLIOTT PROCESS INSTRUMENTS

Century Works, Connington Road,

Lewisham, London SE13 7LN

TV Studio Engineer

The Road Transport Industry Training Board has in operation at its Wembley Headquarters a 3 camera broadcast-quality colour television studio with full telecine and video recording facilities which includes R.C.A. TR50 and 1" Helical Scan systems. We now wish to appoint an experienced studio engineer to join a small team working on the production of training and educational television programmes. The applicant should not be less than 24 years of age and have a good working knowledge of the above equipment.

The starting salary will be in the region of £4,000 depending on qualifications and experience; other benefits include four weeks' holiday, contributory pension and life assurance scheme.

Please send relevant personal history stating how the above requirements are met, and quoting reference ZH.462, to:



Mrs. H. M. Brown, Personnel Manager
 Road Transport Industry Training Board
 Capitol House, Empire Way
 Wembley, Middlesex HA9 0NG



How do you value your electronics testing ability?

All the benefits of the world's largest radio-telephone exporters, could soon be yours. If you value your expertise highly, this is where to get most mileage from it.

Career progression paths are long and wide—the Company's expansion rate has been unaffected by the present economic situation. Equally significant is the importance Pye Telecom rightly attaches to fault-finding and testing to exacting specifications their VHF/UHF advanced design communications equipment. Reasons are obvious — not only is reliability crucial in furthering the Company's progress, but frequently lives depend on the performance of the equipment, because fire, police and ambulance services use it extensively.

So if you have practical experience of this work, maybe in the armed forces, it will pay you handsomely to get full information about the conditions, the relocation assistance and other attractions. Work and live in the attractive university city of Cambridge: alternatively in the nearby expanding town of Haverhill where there are excellent possibilities for private and rented housing.

Phone or write today to

Mrs Audrey Darkin
Pye Telecommunications Ltd
Cambridge Works
Elizabeth Way, Cambridge CB4 1DW
Tel: Cambridge 58985

or

Mrs Cath Dawe
Pye Telecommunications Ltd
Colne Valley Road
Haverhill, Suffolk CB9 8DU
Tel: Haverhill 4422



A member of the Pye, Cambridge Group

(4879)

UNIVERSITY OF EXETER

ELECTRONICS TECHNICIAN GRADE 5

The University of Exeter has a vacancy in the Department of Biological Sciences for an Electronics Technician, Grade 5. Duties include the servicing and repair of a wide range of electronic analytical instrumentation, and the design, construction and modification of electronic units concerned with instruments.

Minimum qualifications: O.N.C. or equivalent. Salary £2,439-£2,895 p.a.

Applications to: The Secretary, University of Exeter, Northcote House, The Queen's Drive, Exeter, Devon. Closing date: 1st September, 1975.

Please quote reference no. 1/75/5055.

(4888)

CHELSEA COLLEGE University of London

TECHNICIAN GRADE 5

required to be responsible for the running and technical development work of Physics and Electronics Undergraduate Teaching Laboratories in Chelsea, SW3. Salary £2849-£3305 per annum, including London Allowance.

Further details and application form from Mr. M. E. Cane (5 PET), Chelsea College, Department of Physics and Electronics, Pulton Place, London SW6 5PR.

4872

ELECTRONICS TECHNICIAN aged 20-30. We are a small Company situated in S.W. London. We require a technician to join our young electronics team working on professional audio equipment. He will be responsible for the alignment, testing and maintenance of digital, analogue and audio circuitry, and should have some practical experience in one of these fields. The company operates a profit sharing scheme. Telephone Mr. Evans at 01-542 1471.

(4869)

CHELSEA COLLEGE UNIVERSITY OF LONDON

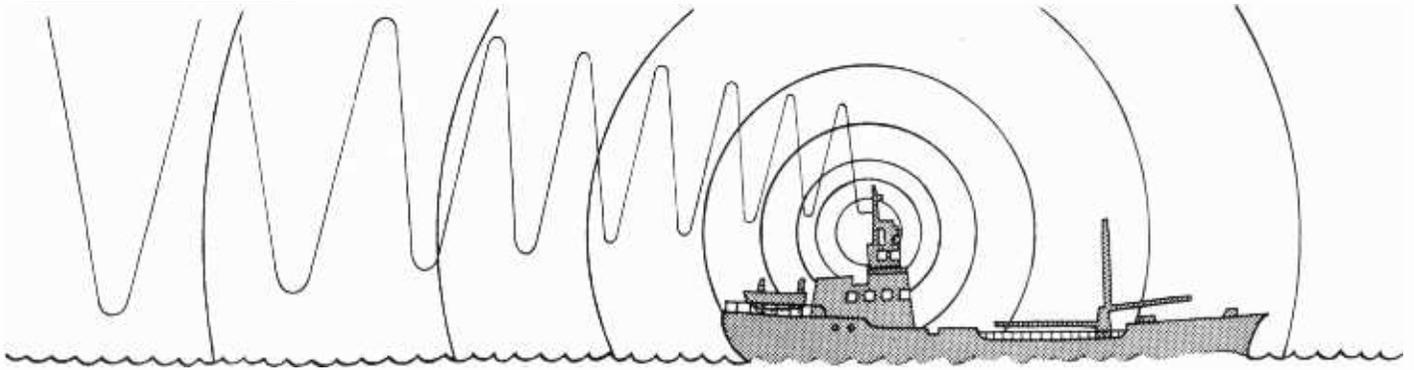
ELECTRONICS/TECHNICIAN

required to take charge of Electronics Workshop for the design and production of prototype electronic equipment for electronics and physics research and teaching, and also for the servicing and maintenance of a wide range of commercial electronic equipment. A wide practical experience and a sound theoretical knowledge of electronics is essential. Experience in microwave instrumentation would be an advantage. 5-day 37½-hour week. Salary (Grade 6) £3,254-£3,860 p.a. including London Allowance. Further details and application form from Mr. M. E. Cane (EW6), Chelsea College (University of London), Departments of Electronics and Physics, Pulton Place, Fulham, London SW6 5PR.

(4886)

TELEVISION TECHNICIANS wanted for Middle East position. Five years heavy maintenance required. Send resumes and copies of certificates to Box WW 4851.

STUDIO IN KENSINGTON AREA requires JUNIOR TECHNICIAN (18-22 years) to assist with maintenance and tape editing. Contact Graham Stephens, 108 Cromwell Road, London SW7. Tel. 01-370 1442. (4887)



Radio Officers—now you can enjoy the comforts of home.

Working for the Post Office Maritime Services really makes sense. You still do the work that interests you, but with all the advantages of a shore-based job: more time to enjoy home life, job security and good money. To qualify, you need a United Kingdom Maritime Radiocommunication Operator's General Certificate or First Class Certificate of competence in Radiotelegraphy, or an equivalent certificate issued by a Commonwealth Administration or the Irish Republic.

Starting salaries, at 25 or over, are £2905 rising to £3704 after three years service. Between 19 and 24, the starting salary varies from £2234 to £2627

according to age. You'll also receive an allowance for shift duties which at the maximum of the scale averages £900 a year and there are opportunities to earn overtime. There's a good pension scheme, sick pay benefits and prospects of promotion to senior management.

Right now we have vacancies at some of our coastal radio stations, so if you're 19 or over, write to: ETE Maritime Radio Services Division (R/B/9), ET 17.1.1.2., Room 643, Union House, St. Martins-le-Grand, London EC1A 1AR.

Post Office Telecommunications

93

ENGINEER

£2500-£3100 + car allowance

The man appointed will be engaged principally on the maintenance of ITEL automatic typewriters both at our premises and in the field. Applicants should therefore have a working knowledge of the IBM Selectric, and should live in the S.E. London/Kent area.

It is hoped, however, that someone with a wide-ranging interest in electronics and able to work on a variety of other projects will be appointed.

It is anticipated that the post might appeal to a young man to whom an informal and flexible environment is not a disincentive.

Phone: Raymond ffoulkes (Godstone 3106) for details, or write naming two referees:

COMPUTER APPRECIATION
Castle Street, Bletchingley
Surrey, RH1 4NX

4896



Opportunities in the ELECTRONICS FIELD

Men with analogue or digital qualifications / experience seeking higher paid posts in: TEST - SERVICE - DESIGN - SALES.

Phone: Mike Gernat, Ref. W.W.

NEWMAN APPOINTMENTS
360 Oxford Street, W.1,
01-629 0501

(94)

LABORATORY TECHNICIANS

Equipment Dept.,
Chiswick, W4.

Interesting work testing new electronic equipment made by the BBC for its colour television and stereo radio services, involving analogue and digital techniques over a frequency range from D.C. to U.H.F.

Qualifications O.N.D., O.N.C. or C. & G. Part II in Telecommunications or Electrical Technician certificate. Initial salary range normally £2,127 to £2,319 rising to £2,952. Good opportunities for promotion to Senior Laboratory and Engineering Technicians with salaries rising to £3,762. Further promotion to Engineer grades is also possible.

Staff will be based at Equipment Department, Chiswick which is within easy reach of British Rail and London Transport services and the M4, North and South Circular roads.

Good club and canteen facilities are available.

The posts are pensionable with four weeks' leave annually.

Requests for application forms to The Engineering Recruitment Officer, BBC, Broadcasting House, London W1A 1AA, quoting reference 75.E.4047/WW. Please enclose an addressed envelope at least 9" x 4" with your application; no stamp required. Closing date for completed application forms is 14 days after publication.

(4877)

WIRELESS TECHNICIANS

There are vacancies at Home Office Wireless Depots throughout England and Wales for Wireless Technicians to assist with the installation and maintenance of VHF and UHF Systems. Ability to drive a car and possession of a current driving licence is desirable.

Salary

is £2010 (at 17), £2450 (at 21) and £2905 (at 25) rising to £3385 a year.

A London Weighting Allowance of up to £410 a year is also payable for staff in London.

A Secure Future

with a non-contributory pension scheme, good prospects of promotion and a generous leave allowance. There are opportunities for day release to obtain higher qualifications.

Qualifications

Candidates should have good experience in Telecommunications and preferably hold a City and Guilds Intermediate Telecommunications Certificate or equivalent.

Interested?

Then write or telephone for further details and application form to:

Mr. C. B. Constable,
Directorate of Telecommunications,
Home Office,
60 Rochester Row,
LONDON SW1P 1JX.
Telephone No. 01-828 9848 (Extension 734).

4877



Home Office

TECHNICIAN TRAINEES

Intelligent practical young school leavers offered opportunities to train ultimately as Public address and sound recording Engineers. Day release scheme, must be of smart appearance and live with parents in Central London area. Write or telephone for interview to:—

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Tel. 01-499 1231

(4832)

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requires

Licensed Helicopter Engineers. Applicants must possess current C.A.A. Licences for Bell 212 and 206B. Attractive Salary / Fringe benefits. Please apply in confidence to:-

The Personnel Officer
Sabah Air
P.O. Box 747
Kota Kinabalu
Sabah, Malaysia

(4893)

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Applicants should have HND or better qualifications, preferably with some Industrial Design experience in transformers from five to several hundred kVA and/or industrial semi-conductor rectifier equipment. Training in the design of the Company's specific products will be given.

Salary commensurate with qualifications and experience and four weeks annual holiday.

Assistance will be given with removal expenses where appropriate.

Please apply in writing to:

Mrs J. Davey, Personnel Services

(4841)

BRENTFORD ELECTRIC LIMITED

Manor Royal, Crawley RH10 2QF - Tel: Crawley 27755



ELECTRICITY ENGINEERING DEPARTMENT COMMUNICATIONS AND MEASUREMENT FIRST ENGINEER—DESIGN AND APPLICATION

A First Engineer is required for duties in the Technical Section of the Chief Engineer's Department at Manweb Head Office, Sealand Road, Chester.

He should have a sound theoretical knowledge and extensive practical experience in the communications field.

The post involves work on projects in the following branches of communications:

- (1) AUTOMATIC TELEPHONY
- (2) POST OFFICE TELEPHONY
- (3) PRIVATE RADIO SYSTEMS
- (4) DATA TRANSMISSION
- (5) CARRIER LINE TRANSMISSION
- (6) COMPUTER PRINCIPLES AND APPLICATIONS

(A) The First Engineer will be required to share responsibility for the design of extensions to the communications systems and of new applications of computer technology.

(B) To negotiate on engineering design matters with Post Office engineers and manufacturers of communication equipment, and must be able to prepare technical specifications of communication equipment

(C) And to control development work at a technical workshop close to Head Office and also at a computer terminal located at Head Office and provided with communication links to the Manweb computers and to the private communications systems.

Manweb is operating an extensive communications system linking district offices and depots at the ten districts with Head Office, and the first stage in the provision of links between districts and the Head Office computers has been implemented.

Salary within the range £4630-£6680 per annum + £120 responsibility payment + £229.35 threshold payment

Applications, giving full details, should be sent to the Secretary (Personnel), Manweb, Head Office, Sealand Road, Chester, to arrive not later than 8th September, 1975 (4840)

UNIVERSITY OF SHEFFIELD RESEARCH TECHNICIAN (Grade 3-5)

required for the Space Physics Group within the Department of Physics for an initial period of 2 years. The successful candidate would be primarily concerned with the research, development and construction of pay loads for use with ionospheric sounding rockets. Experience of design and/or construction in one or more of the following areas would be advantageous:

- (a) low noise analogue circuitry D.C. - 100 KHz
- (b) Radio frequency circuitry 100 MHz - 1500 MHz and modulation methods
- (c) Ultra reliable equipment for use in extreme environments and/or prolonged periods of unattended operation.

A current driving licence is essential, the duties may include some travel both within the U.K. and abroad for periods up to several weeks.

Salary scales £2013-£2343 p.a. or £2247-£2628 p.a. or £2439-£2895 p.a.

Please write to the Deputy Director of Services (Ref. S356/WW), The University, Sheffield S10 2TN. (4844)

VALVES WANTED

WE BUY new valves, transistors and clean new components, large or small quantities, all details, quotation by return. — Walton's, 55 Worcester St., Wolverhampton. (62)

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GLASGOW. Hi Fi, Cassette Decks, Tape Recorders, Video Equipment, always available we buy, sell and exchange for Hi-Fi sets and photographic equipment. VICTOR MORRIS Audio Visual Ltd, 340 Argyle Street, Glasgow, G1, 8/10 Glassford Street, Glasgow, G2, 31 Sauchiehall Street, Tele: 041-221 8958. (11)

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DUBILIER CAPS .02 12KV, .01 15KV, £1.50; .05 4KV, .1 3KV, .01 10KV £1; .05 2KV, .0144KV, .001 10KV, 60p. P.p. 15p. G. Radio Barrel W'meter, 55-400Mc £5. Old 4 1/2" meters 1/2 MA £1. Old 12V po relays 2 for £1. Potmalt, CU337 £4. C.W.O. Plasma. Elects. PO Box 59, Uxbridge. (4892)

STODDART RADIO INTERFERENCE MEASURING SET. 3 Units range from 15m/cs to 1000mc/s. In perfect working order. £2,500.00 Hewlett Packard Spectrum Analyser Type 8551A. Not working on suitable for spares £400.00. Telephone evenings only 01-449 6548. (4881)

Natural Environment Research Council British Antarctic Survey

Electronic Engineer (graduate or H.N.C.) required for maintenance, development and operation of radio-echo sounding equipment. This is a pulsed radar system working at 60 MHz used in aircraft to measure ice thickness in the Antarctic. The same equipment is used on the surface to measure glacier flow.

Candidates should be familiar with radar theory and digital logic, and have experience in design and construction in these fields. The successful candidate will be based in Cambridge (Scott Polar Research Institute) and must be prepared to work in the Antarctic for periods of up to four months.

Salary according to age and experience from £2,012. In addition, a Cost of Living Supplement of £229.68 per annum is payable.

Please write for further details, stating full qualifications to:

The Establishment Officer, British Antarctic Survey, 2 All Saints Passage, CAMBRIDGE CB2 3LS.

Tel: Cambridge (0223) 61188

Please quote reference BAS1. 4883

DESIGN/DEVELOPMENT ENGINEER

We are a young but expanding Electronics Company moving to a modern well-equipped factory. An opening has arisen for an Engineer to join our research and development department

The successful applicant will be involved in all sides of project engineering and should have experience in both analogue and digital electronics

The job offers an exciting future with rapid advancement to Senior Management for the right person

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Apply Technical Director
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BR5 3QW (4890)

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LABORATORY and Test Equipment. Oscilloscopes—Cossor CDU 150 DB 35MHz £225. Solartrons. CD1212 DB 24MHz £90. CD523S2. SB 10MHz £45. CD711S2 DB 10MHz with Xtal marker £40. CD513.2 SB £37. CD513 SB £25. Airmec 723 £10. SIG GENS. CT373, inc VVM and distortion meter, 17Hz -170KHz £65. Solartron OS101. 25Hz-250KHz £24. Sullivan RC 40Hz-125KHz £20. CT378 Avo 2-220MHz £25. Avo 50KHz-80MHz £20. Marconi TF144G 85KHz-25MHz £25. Pulse 0.5us-1600us £15. CT202 sweep 7-70MHz £15. Marconi TF1073A sweep 10KHz -30MHz £15. CT432 Xtal 100KHz 1MHz 10MHz and ext. £7.50. Marconi Saunders 1.3-4 2GHz £95. All plus VAT 8%. Amperon Ltd., 39 Kent Road, Dartford, Kent. Tel. 20433. (4891)

MIKES WANTED. Old ribbon mike Marconi/EMI type ABXT with stand, working condition, also other pre-war mikes. Mr K. E. Eriksson. Radjursvagen 22 S-440 03 FLODA, Sweden. (4892)

VIATRON VDU TERMINAL incorporating Keyboard. 320 character VDU, MOS microprocessor, and twin Philips-type cassette station. £190. VIATRON incremental DATA RECORDER, manufactured 1972. £95. MDS 1320 LINE PRINTER, as new, £1200. FLEXOWRITERS (paper tape typewriters with I/O) from £50. EMI AUDIO Echo Unit, £55. GE Model 661A High Speed incremental paper tape reader (half rack mount) c/w drive & read amps.. £50. NEW paper tape golfball typewriter, £350. HONEYWELL G-115 PROCESSOR Module with 4k store, Power Supply Module, and Control Console (System installed 1971). ALL £165. (Honeywell Series 200 Magtapes (four with controller), £65 per unit or £225 the lot, COMPUTER APPRECIATION, Godstone 3106, Otford 3256. (4897)

ANNOUNCEMENT

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Due to GOVERNMENT V.A.T. REGULATIONS the cost of the additional office work involved in applying our 2½% discount is prohibitive. We have no option but to reluctantly withdraw our discount for the present. This does not reflect our policy but is directly due to GOVERNMENT REGULATIONS. We hope you will join with us in looking forward to better times and to a more sane fiscal policy.

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Tel: West Kingsdown 2344

APPOINTMENTS

AGRICULTURAL RESEARCH COUNCIL WEED RESEARCH ORGANISATION ELECTRONICS ENGINEER

This new post involves preventive maintenance, repair and development of new electronic devices. Equipment involved includes electronic aspects of controlled environment cabinets and rooms, monitoring instrumentation, general laboratory equipment, and data collection and handling facilities. The engineer will work closely with research staff to apply the potentialities of contemporary electronics to a wide range of research activities. Hence, in addition to general proficiency in electronics, a broad knowledge, inventive ability and capability to communicate with other specialists will all be required.

Minimum qualifications required are ONC or equivalent qualification in Electronics Engineering.

The appointment will be in the grade of Professional and Technology Officer III (£3,450 x 5 to £3,925).

Non-contributory pension scheme.

For further particulars and application form write to The Secretary, Weed Research Organisation, Begbroke Hill Yarnton, Oxford OX5 1PF, quoting 7/75. Closing date for applications 10th September, 1975.

(4875)

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To: Mr. J. Bandeen
Executive Director (Administration)
G.R. INTERNATIONAL ELECTRONICS LTD.
Almondbank, Perthshire PH1 3NQ

(4904)

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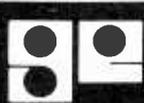
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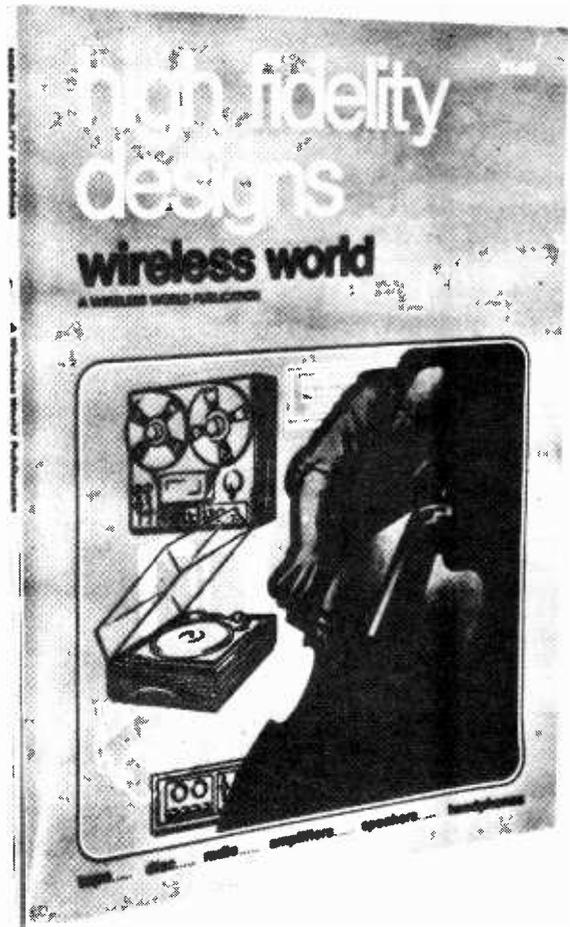
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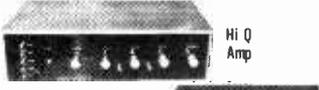
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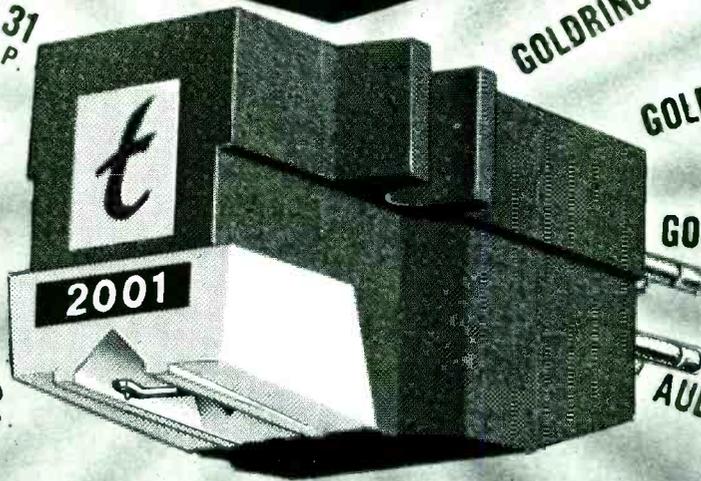
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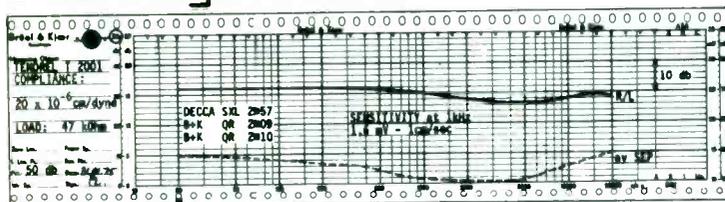
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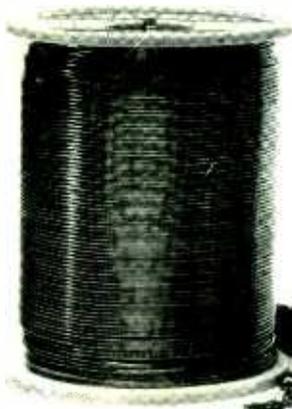
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304D } 304W }	mildly activated Halide Free	10% } 25% }	MIL-F-14256D Type RMA; DTD 599A
PC. 21A	activated	38%	DIN 8527 Type F-SW 32
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