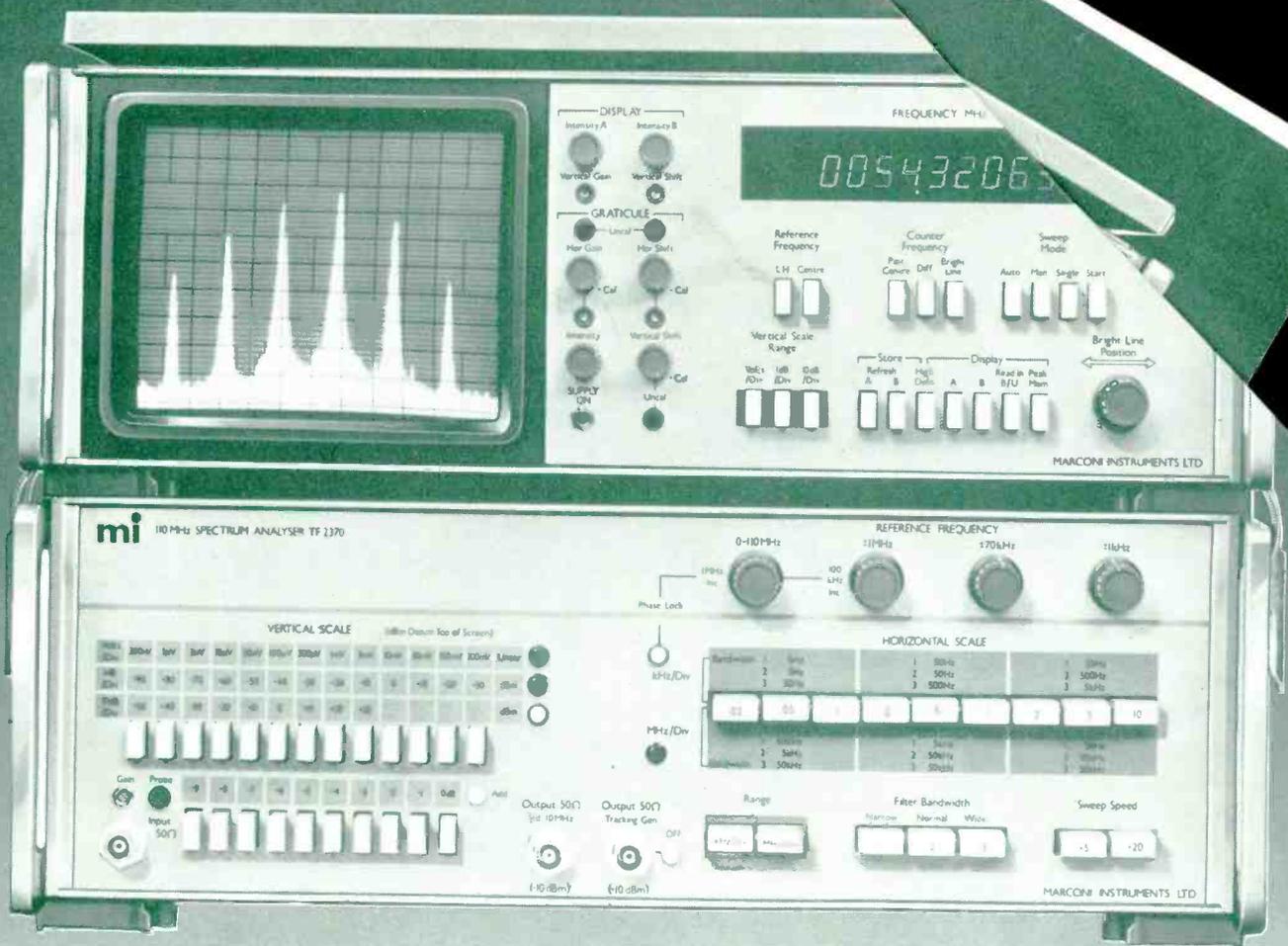


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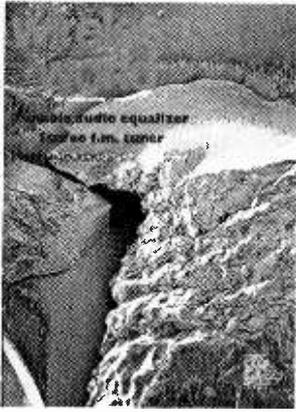
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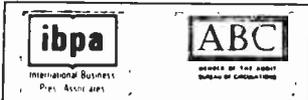
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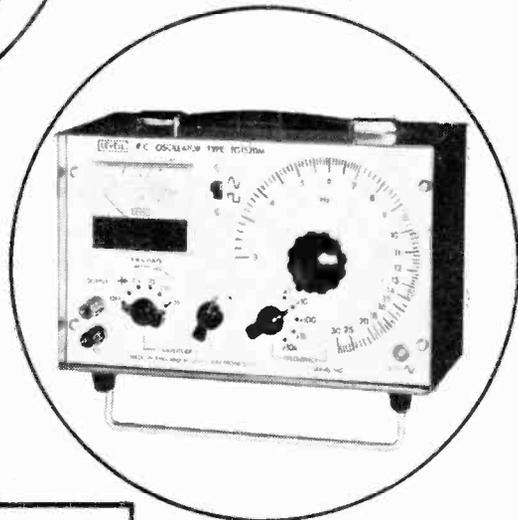
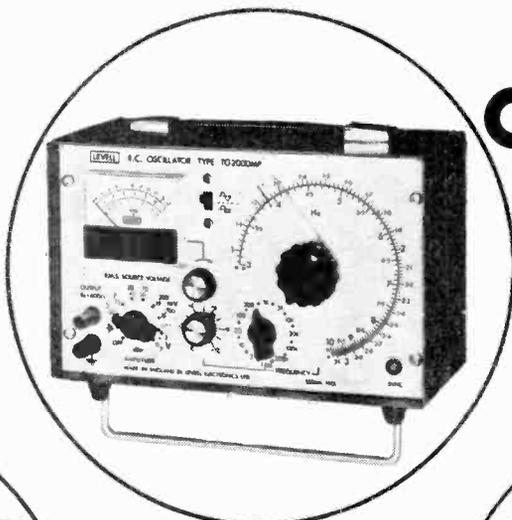
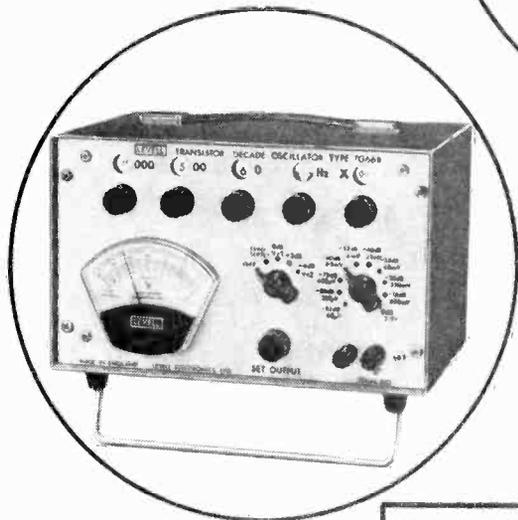
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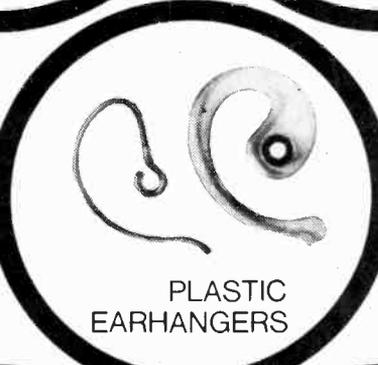
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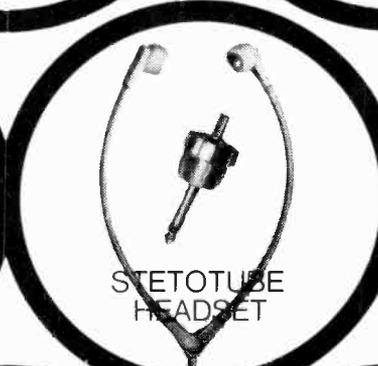
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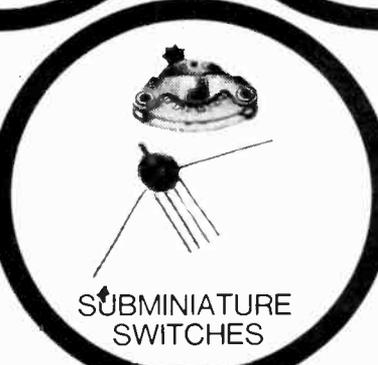
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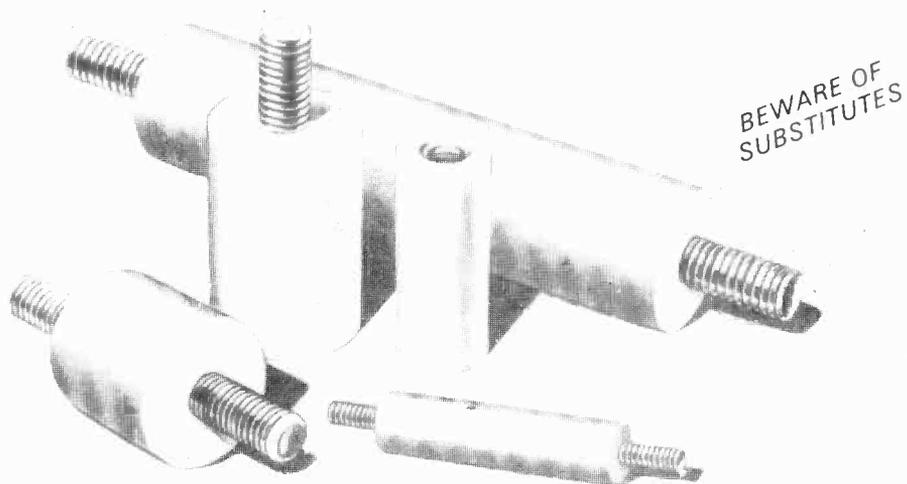
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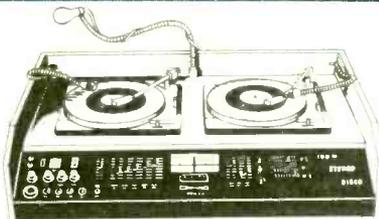


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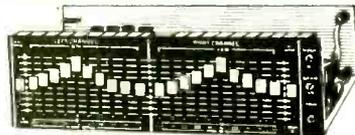
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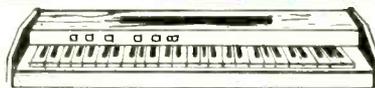
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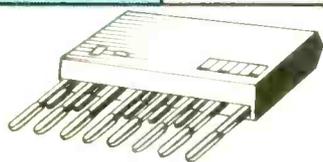
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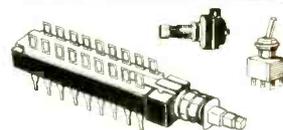
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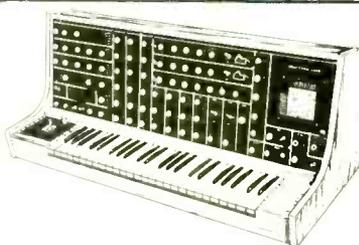
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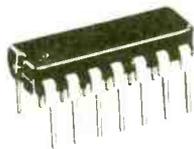
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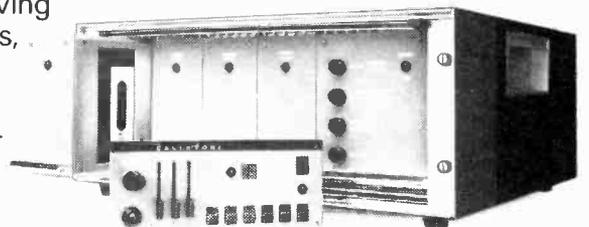
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www.americanradiohistory.com

JVC's new portable 3-tube CY-8800E camera

To be unveiled at the
International Broadcast
Convention

Featuring a choice of $2/3$ " Plumbicon or Saticon tubes, the new JVC portable 3-tube camera is one of the latest products of the company's 21 years' involvement in video systems. Pictures through the 10:1 f/1.9 C mount zoom lens, fitted as standard, look crisp and clean with better than 500 lines horizontal centre resolution.

Designed for Europe, the new PAL CY8800E has a wide range of facilities to enable operators to record virtually any scene – indoors or on location. With a 49dB S/N ratio at f/4 and only 3000 lux illumination, the camera can be used in light levels as low as 300 lux.

Standard features include built-in colour bar generator, $1\frac{1}{2}$ " viewfinder with a 3" magnifier, 3 way sensitivity switch (normal, +6dB, +12dB), auto white balance and bias lighting when the camera is fitted with Saticon tubes. Weight is only 9.6kg including lens, and viewfinder.

Applications for this superb camera

Its inherent flexibility and sensitivity make this JVC portable camera ideal for almost any situation, including training, news-gathering, event recording, pilot commercial preparations and sport.

More new JVC units

To complement the CY-8800E camera, JVC is also introducing a gen-lock unit (GN-8800E), 3" high-resolution viewfinder (VF-8300E), a more sophisticated electronic editing suite (CR-8500E), an automatic editing control unit (RM-85E), and a remote and sync unit (RS-8800E) for CY8800E studio applications.

Together, this equipment adds up to the world's finest $3/4$ " U-format video system.

For more information about all this new JVC video equipment, use the inquiry service to obtain literature and a list of Bell & Howell video dealers.



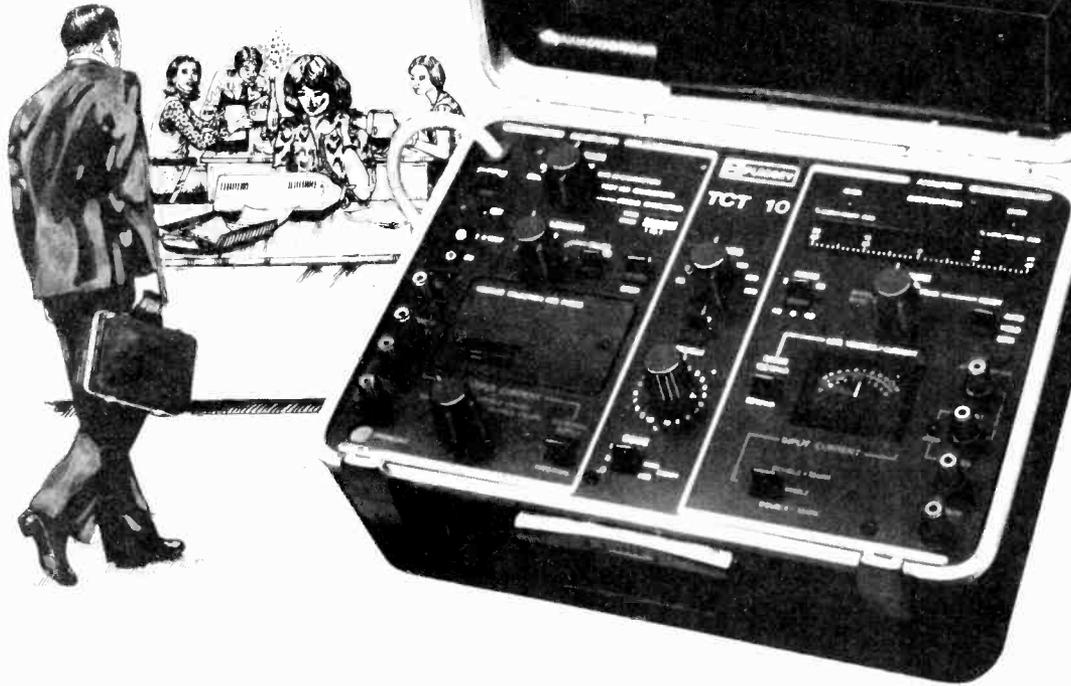
BELL & HOWELL

Bell & Howell A-V Ltd.,
Alperton House, Bridgewater Road,
Wembley, Middlesex HA0 1EG.

WW-090 FOR FURTHER DETAILS

Telegraph testing simplified

TCT10



Lightweight, portable, Telegdata TCT10 makes light work of on-site circuits and machines.

This new Plessey instrument combines signal generator and analyser in a single briefcase-size unit enabling on-site testing of telegraph circuits and machines to be carried out speedily and with a high degree of accuracy.

Powered from the a.c. mains supply, the TCT10 gives a choice of output levels and test signals in CCITT No 2 and No 5 alphabets including the full 96 character 'fox' message, Q9S and any single character on demand.

Accurate readout (to 1%) is given unambiguously on an LED scale registering up to 40% distortion early/mark bias and late/space bias.

Full specification is available in a colour-illustrated brochure. See how your telegraph test operations can be improved — telex or write to: Telegdata Department, Plessey Controls Limited, Sopers Lane, Poole, Dorset, United Kingdom BH17 7ER. Telex: 41272.



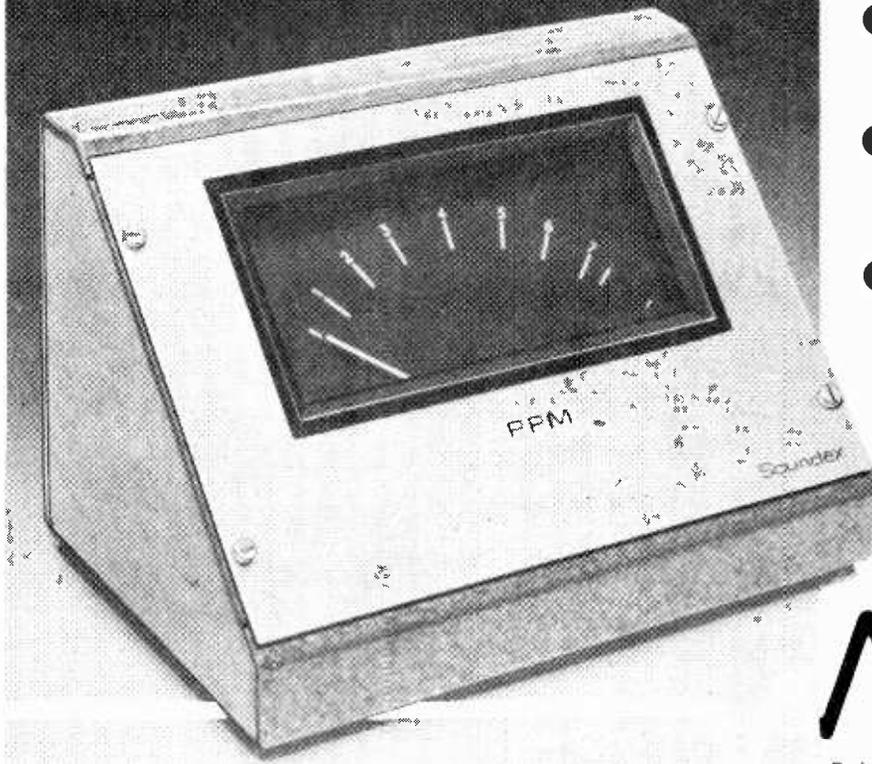
**PLESSEY
CONTROLS**

WW—057 FOR FURTHER DETAILS

600-2-224

The Soundex PPM

COMPLIES WITH B.S. 4297 AND IEC 468



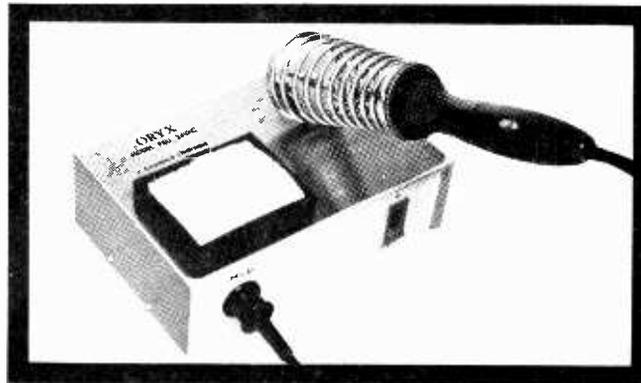
- ROBUST, FREE-STANDING UNIT
- 24V DC POWER SUPPLY AVAILABLE
- ILLUMINATED PPM OR dB SCALE
- EACH UNIT SUPPLIED WITH INDIVIDUAL CALIBRATION CERTIFICATE
- FREQUENCY RESPONSE FROM 15Hz TO 35KHz

Bulgin Electronics

One of the Bulgin Group of Companies

Park Lane, Broxbourne, Hertfordshire. Tel: Hoddesdon 64455

WW-119 FOR FURTHER DETAILS



Introducing the **ORYX PSU 24**

a new compact self-contained 24 volt power supply unit
for ORYX temperature controlled soldering irons.

Styled in tough plastic, the ORYX PSU 24 is a smart new supply unit that is self-contained and small enough for the smallest of benches. Designed to meet BSS 3456 the ORYX has all the features you need - and more:- ON/OFF illuminated rocker switch; 3 pin non-reversible socket supplying 24 volts; a BSS 3535 transformer; an outside primary fuse; 1.5 metre white cable to BSS 6500 and fuse protection for transformer secondary wiring.

A unique feature is the facility to modify a 3 wire power system to a 2 wire fully isolated unit and vice versa.

A new product from Greenwood Electronics

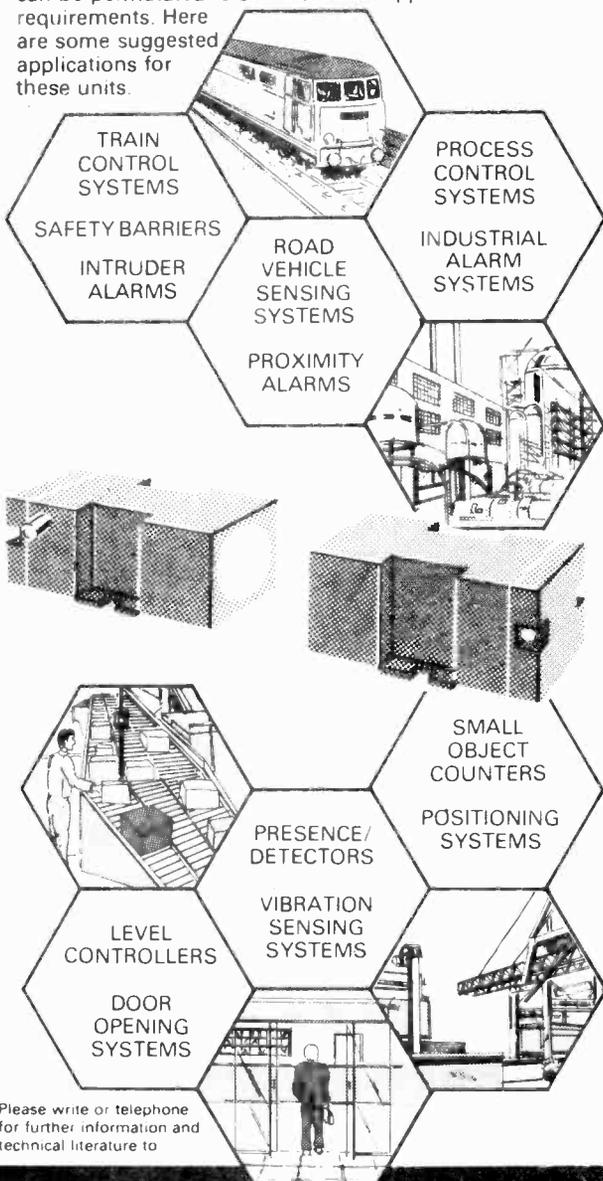
For full technical data write for information to Greenwood Electronics, Portman Road, Reading, RG3 1NE Telephone. Reading (0734) 595844 Telex 848659

WW-075 FOR FURTHER DETAILS

THE JAMES SCOTT INDUSTRIAL MICROWAVE RANGE OF EQUIPMENT

The James Scott range of Microwave equipment offers industrial users a greater choice of alternative systems in robust, industrial, cast aluminium housings, for a wide variety of applications.

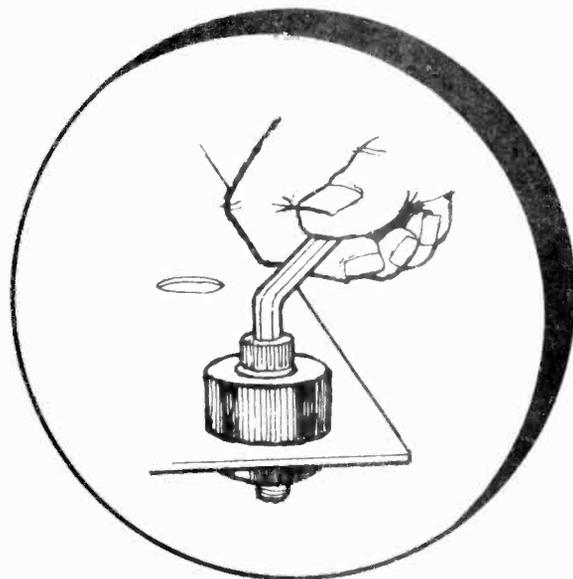
The range is made up of standard sub-assemblies which can be permuted to suit individual application requirements. Here are some suggested applications for these units.



Please write or telephone for further information and technical literature to

JAMES SCOTT
(Electronic Engineering) Ltd
CARNTYNE INDUSTRIAL ESTATE
GLASGOW G32 6AB Tel: 041 778 4206 Telex 778286

WW-025 FOR FURTHER DETAILS



FOR CLEAN HOLES FAST!

- Easiest and quickest way of punching holes in sheet metal (up to 1.625mm mild steel)
- Q-MAX stands for quality and reliability
- Holes are punched cleanly and no filing is necessary
- Continuous even load during punching
No jagged edges. Burr free hole
- Specially heat treated to maintain keen cutting edge
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- Simple operation, saving time and energy

57 METRIC & LINEAR SIZES



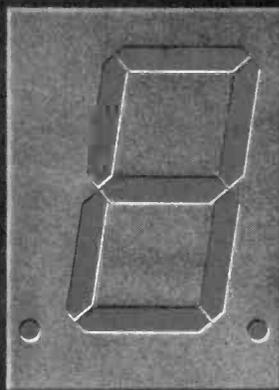
SHEET METAL PUNCHES

Q-Max (Electronics) Ltd ^(R)

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London EC4A 1JQ
Telephone: 01-242 7400

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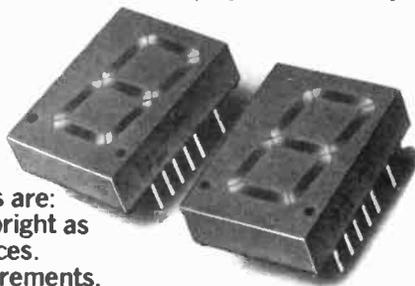
FOR CLARITY
YOU'D HAVE TO GO
A LONG WAY
TO BEAT US.



Hewlett-Packard's HDSP-3400 series display.

The readability of the HDSP-3400 series is excellent—designed for viewing distances of up to 33 feet. Even in bright ambients the clarity is remarkable. They replace existing 0.6" and 0.8" displays and are fully compatible with them.

For full details of why you'd have to go a long way to beat the HDSP-3400 series return the coupon.



Significant features are:

- * Nearly twice as bright as competitive devices.
- * Low power requirements.
- Single Ga AsP chip per segment.
- * Categorized for luminous intensity, assuring uniformity of light output.
- * Grey packaging for optimum contrast.
- * Overflow digit device available.
- * Low cost. 100 + £1.36.



HEWLETT  PACKARD

To Crellon Electronics Ltd., 380 Bath Road, Slough, Berks. SL1 6JE.
Tel: Burnham (06286) 4434. Telex: 847 571.

Please send me full data on the Hewlett-Packard HDSP-3400 Series

Name _____

Company _____

Address _____

Tel. _____

WW9

WW-009 FOR FURTHER DETAILS

ELECTRON

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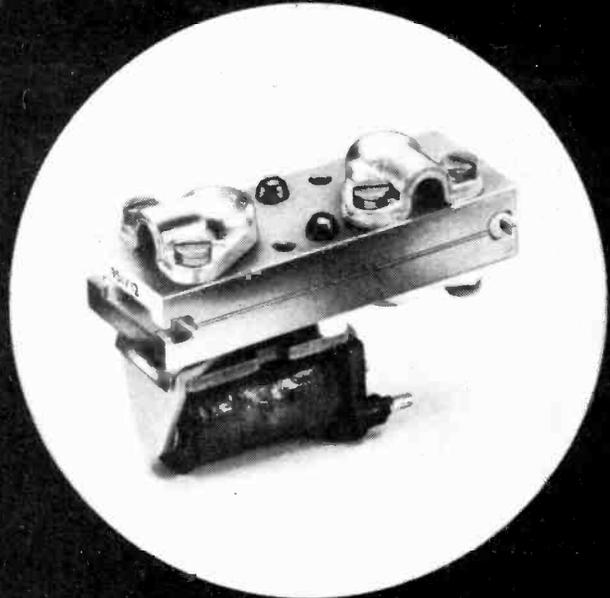
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Coaxial Relay



for aerial switching

Pye Electro-Devices' Series 951 coaxial relay is designed for high speed switching of signals up to 450 MHz. It can be used to advantage in applications requiring exceptionally low inter-contact capacitance.

Multi-position clamps enable UR43 type coaxial cables to be connected directly to the relay body. Connection is simple and efficient. And there's no need for coaxial plugs and sockets. Consequently the relay is extremely compact, and weighs less than 100g.

The contacts are rated at 1A or 50W maximum, and coils are available for operation at up to 100V d.c.

**Contact us today
for further information.**



Pye Electro-Devices Ltd.

Controls Division, Exning Road, Newmarket, Suffolk CB8 0AX
Tel: Newmarket (0638) 5161 Telex: 81245



Artistic licence?

We at QUAD go to a very great deal of trouble to ensure that with a QUAD 33 in the Cancel position, the voltage delivered to your loudspeakers is a virtually exact RIAA transfer of the voltage the pickup will produce into a stated passive load. Nothing added – nothing taken away.

A visiting journalist recently suggested that we should not do this. Final adjustment should be done by ear, he said.

What an opportunity!

After all we know that if we add a little warmth with a subtle boost in the lower middle and balance this with an ever so gentle hump in the quack region (2-3kHz), we can make most programmes sound superficially more impressive. Come to that, why not change the 3180 μ S to 5000 μ S adding a little more 'heft' that most people will fall for. We could even make a special model for the boom and tizz brigade.

Been to any live concerts recently ?

For further details on the full range of QUAD products write to :
The Acoustical Manufacturing Co. Ltd.,
Huntingdon, Cambs. PE18 7DB
Telephone : (0480) 52561

QUAD



for the closest approach to the original sound
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For high quality electronic valves, semiconductors and integrated circuits — and the speediest service — specify Haltron. It's the first choice of Governments and many other users throughout the world. Haltron product quality and reliability are clearly confirmed. The product range is very, very wide. And Haltron export expertise will surely meet your requirements. Wherever you are, get the best service. From Haltron.



Hall Electric Limited,
Electron House,
Cray Avenue, St. Mary Cray,
Orpington, Kent BR5 3QJ.
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Telex: 896141

WW—029 FOR FURTHER DETAILS

SME SERIES III

The best pick-up arm in the world



**Design Council
Award 1978**

'This new pick-up arm is the result of in-depth study of disc reproduction requirements and reflects the audio perfectionist's attitude to design which we have come to associate with SME'

'What really matters is that the manufacturer has anticipated cartridge developments and this alone puts the component well ahead of most others.'

What Hi-Fi?
'... judging by the results obtained from the moving-coils I would say that with damping, this arm is very good and gives a high performance. Certainly it is well in advance of the S2 and now, in the eyes of the establishment, will presumably become the best pick-up arm in the world.'

Hi-Fi Answers
'There is no doubt that this is an exciting arm well in the avant garde of today's audio electronics. It will certainly have an irresistible attraction for the enthusiast: its precision engineering and its conspicuous aid to music listening will extend this appeal to music lovers.'

Financial Review (Australia)
'The law of diminishing returns says the Series II Improved is a better buy, the law of perfection says that the Series III is a better engineered and more universal arm. For many people that will settle the argument.'

Electronics Today, International (Australia)
'... the SME Series III must surely rank with the top few arms, having as an overwhelming benefit tremendous flexibility of use.'

Hi-Fi Review (Australia)
'For our examples we chose the new SME Series III arm. It could hardly be more appropriate, for SME—a British firm with an enviable reputation for getting things right—are noted for attention to the finer points which we advocate to hi-fi buyers.'

Popular Hi-Fi

'Starting from scratch, or if you are obsessed with perfection, the SME Series III metriculously set up must have the edge, probably not only over their earlier design but also over that of any other.'

The Gramophone
'... the new SME arm fulfils a lot more of the criteria mentioned than most arms, it was gratifying to hear the results achieved with Pioneer's PLC 590 and the Ortofon SL 20E. The best way of quickly describing the improvement is to say that all the strengths of a moving-coil cartridge were further emphasised, or in other words bass out-put became even more solid and realistic than ever, that incredible open clarity of a good moving-coil design was heightened and importantly, treble quality gained both better definition and sweetness.'

Practical Hi-Fi & Audio
'Standards of engineering in the execution of an SME arm reflects the sort of clear-sighted production competence which caused the British Empire to survive for so long and which continues to earn respect for British high technology overseas. While it takes a true enthusiast to appreciate the finer design points, an SME Series III arm is not only a technological feat it is also a function of a craftsman's love and a work of art. To someone interested more in music than machines and whose aims are high, it should suffice to say that an SME Series III arm is all you could possibly need. It's as simple as that.'

Men Only
'... the new Shure V-15 IV associates for the best with this exceptional arm. In this particular case, this new listening is at the same time more accurate and more human. ... from the cartridges tested the most surprising differences come from Technics and Ortofon MC20 and it is probable that you will have other surprises of the same kind with other cartridges. The arm adapts to every case, but for all the models it brings an improvement of quality, which may vary in proportion, but which always exists.'

Son Hi-Fi Magazine (France)

WW — 014 FOR FURTHER DETAILS

LOW COST · MULTI-CHANNEL · HIGHLY VERSATILE

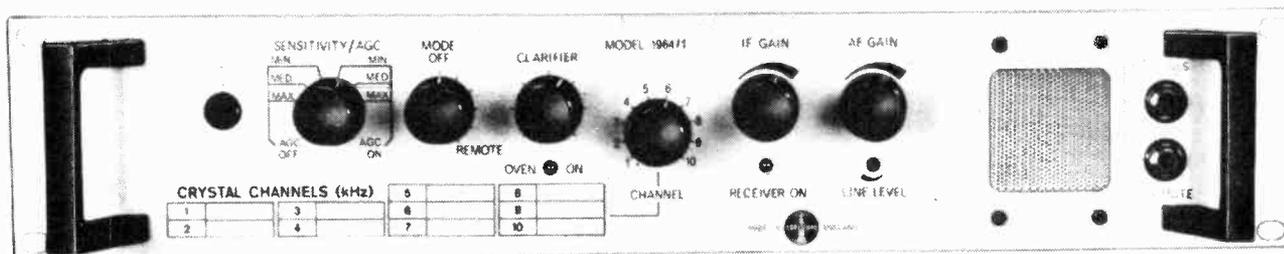
NEW Model
1964
from Eddystone

Eddystone's Model 964 is already in worldwide use for high-performance single channel working. The new multi-channel Model 1964 fulfils a whole new range of communication requirements.

- Pre-determined fixed frequencies
- Standard range 1.6MHz – 27.5MHz
- Versions available 100kHz – 50MHz
- Broadband or tuned RF amplifiers
- Dual diversity operation
- Remotable
- USB: LSB: DSB: F1 telegraphy

Please ask for illustrated brochures giving details of complete range.

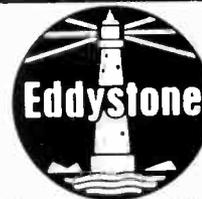
483mm panel to suit standard racking



Eddystone Radio Limited

Member of Marconi Communication Systems Limited

Alvechurch Road Birmingham B31 3PP England
Telephone: 021-475 2231 Telex: 337081
A GEC-Marconi Electronics Company



WW — 067 FOR FURTHER DETAILS

A. D. BAYLISS & SON LTD.

Behind this name there's a lot of real POWER!

Illustrated right is a **TITAN DRILL**

Mounted in a multi-purpose stand. This drill is a powerful tool running on 12v DC at approx. 9000 rpm with a torque of 350 gm. cm. Chuck capacity 3.00 m/m. The multi-purpose stand is robustly constructed of steel and aluminium. The base and bracket are finished in hammer blue.

Also available for use in the stand is the **RELIANT DRILL** which is a smaller version of the Titan. Approx. speed 9000 rpm. 12v DC, torque 35 gm. cm. Capacity 2.4 m/m.

TITAN DRILL & STAND

£21.45
+ 8% VAT = £23.17 + £1 P&P
£9.79 + 8% VAT = £10.57 + 35p P&P

TITAN DRILL ONLY

RELIANT DRILL & STAND

£18.44
+ 8% VAT = £19.92 + £1 P&P
£6.34 + 8% VAT = £6.85 + 35p P&P

RELIANT DRILL ONLY

TITAN MINI DRILL KIT

Drill Plus 20 Tools

£16.25
+ 8% VAT = £17.55 + 50p P&P

RELIANT MINI DRILL KIT

Drill Plus 20 Tools

£13.20
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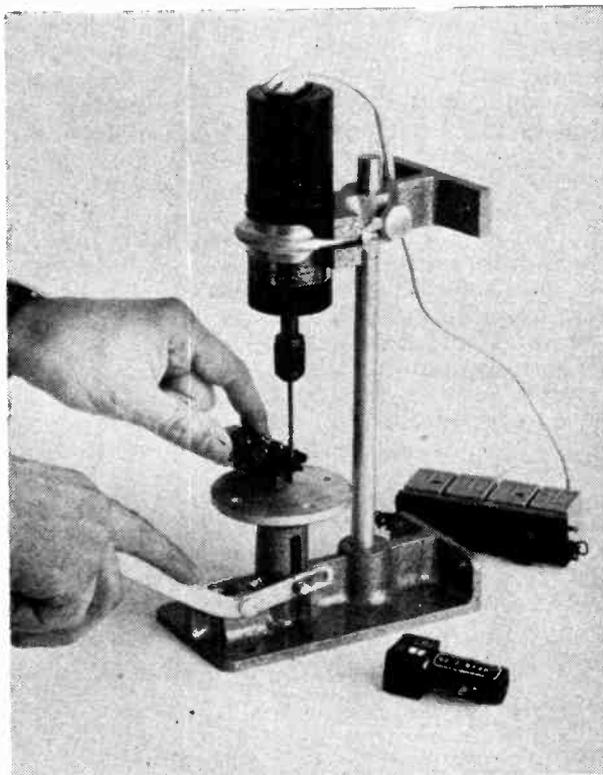
TRANSFORMER UNIT

£9.40
+ 8% VAT = £10.15 + 75p P&P

These are examples of the extensive range of power tools designed to meet the needs of development engineers, laboratory workers, model makers and others requiring small precision production aids.

To back up the power tools, Expo offer a comprehensive selection of Drills, Grinding Points and other tools.

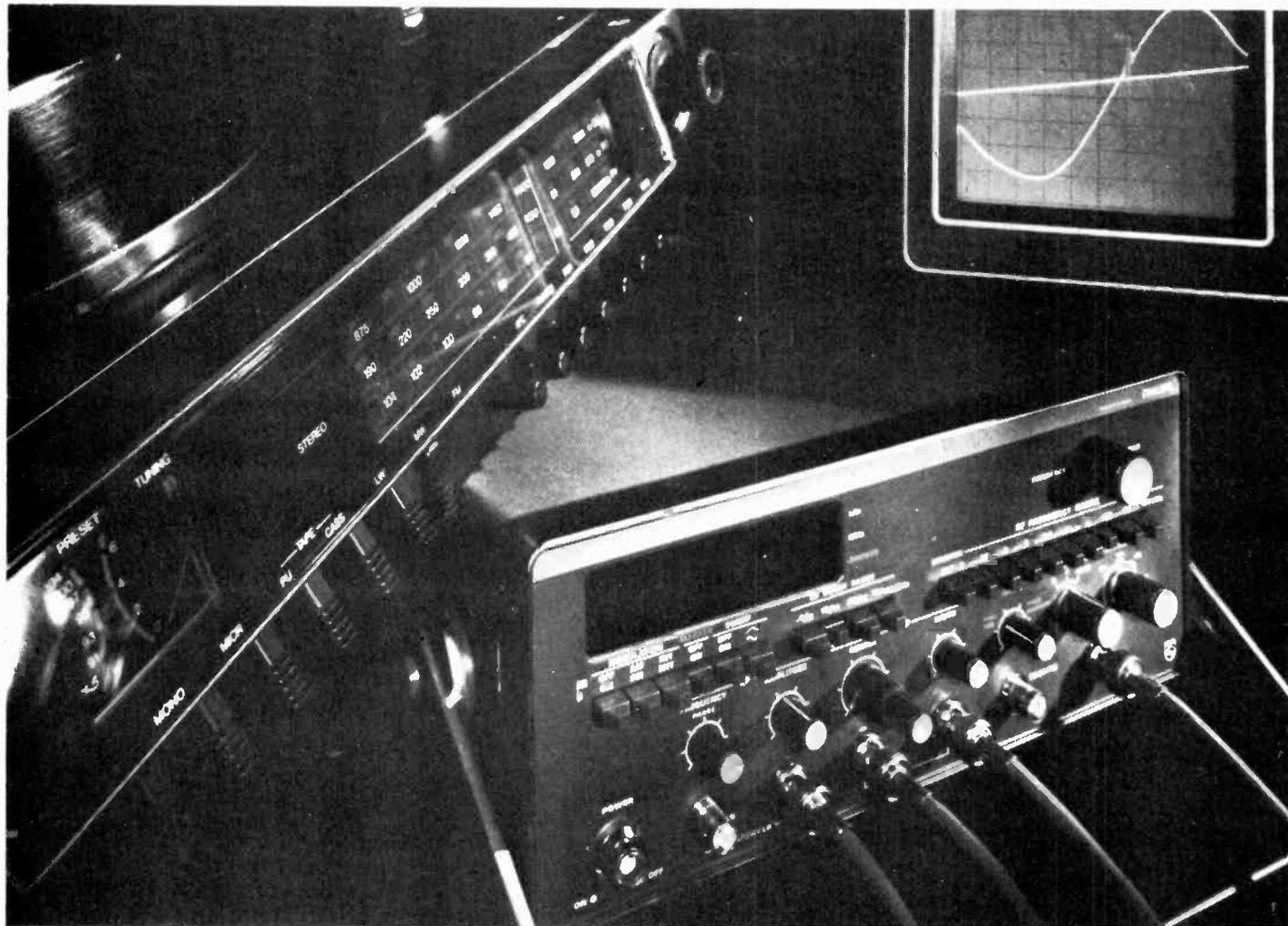
SEND STAMP for full details to main distributors



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Stockists: Richards Electric, Gloucester; Hoopers of Ledbury; Hobbs of Ledbury; D&D Models, Hereford; Bertella, Gloucester; J. Power Services & Co. Ltd., Worcester

WW—037 FOR FURTHER DETAILS



All round performance you can trust

This is the R.F. generator for you

It really is a pleasure to use our R.F. generator PM 5326. Frequency setting must be the easiest yet, press the range button and tune-in over 0.1-125 MHz to one part in 10,000 with the 5-digit LED display. You have to see this to realise how it beats all those dials and multipliers that have been around for some time. Output level is stabilized over all ranges at 50 mV into 75Ω and can be attenuated right down to more than 100 dB!

That means output levels of 0.5 μV are easily available and then only via the RF OUT connector. A "double-box" construction keeps RF radiation very low, enabling you to make accurate sensitivity measurements with full confidence. To these fundamental qualities of precise and stable frequency setting, wide range attenuation and excellent RFI, this generator also has:

- Frequency range 0.1 to 125 MHz
- Internal and external AM/FM modulation facilities
- Four RF sweep ranges for AM/FM IF's, Band II and Video IF's
- Variable and fixed markers
- COUNTER IN connector for checking external oscillators



**Test & Measuring
Instruments**



Pye Unicam Ltd

Philips Electronic Instruments Dept.
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PHILIPS

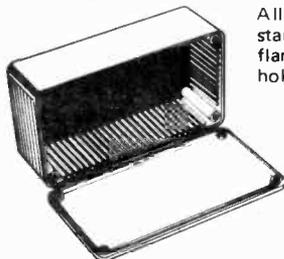
☐ AUDIO AND VIDEO SERVICE EQUIPMENT ☐ AUTOMATIC TEST AND MEASURING EQUIPMENT ☐ COUNTERS AND COUNTER/TIMERS ☐ DC POWER SUPPLIES AND AC STABILIZERS
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WW-094 FOR FURTHER DETAILS

BIMCONSOLES BIMBOXES BIMBOARDS BIMDRILLS BIMDICATORS

ABS & DIECAST BIMBOXES

5 sizes, in either ABS or Diecast Aluminium
ABS moulded in Orange, Blue, Grey or Black
Diecast Aluminium available in Grey Hammertone
or Natural



All boxes incorporate guides on all sides for holding 1.5mm thick pcb's and stand-off bosses in base for supporting small sub-assemblies etc. Close fitting flanged lids held by screws running into integral brass bushes (ABS) or tapped holes (Diecast).

	ABS	Diecast	Hammertone	Natural
(100x50x25mm)	BIM2002/12	£0.95*	BIM5002/12	£1.20*
(112x62x31mm)	BIM2003/13	£1.05*	BIM5003/13	£1.50*
(120x65x40mm)	BIM2004/14	£1.15*	BIM5004/14	£1.86*
(150x80x50mm)	BIM2005/15	£1.30*	BIM5005/15	£2.38*
(190x110x60mm)	BIM2006/16	£2.04*	BIM5006/16	£3.41*

Also available in Grey Polystyrene (112x61x31mm) with no slots and self tapping screws BIM2007/17 £0.88*

MINI DESK BIMCONSOLES

Moulded in Orange, Blue, Black or Grey
ABS and incorporating guides on all sides
for holding 1.5mm thick pcb's. 1mm Grey
Aluminium panel sits recessed into front of
console and held by screws running into
integral brass bushes. Stand-off bosses in
base for supporting small sub-assemblies
etc. 4 self adhesive
rubber feet also in-
cluded.

BIM1005
(161x96x58mm)
£2.12*
BIM1006
(215x130x75mm)
£2.94*



LOW PROFILE BIMCONSOLES



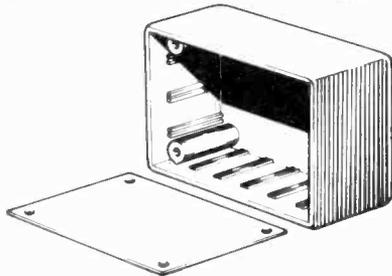
1mm Grey Aluminium panel sits recessed into front of console base, which is moulded in Orange, Blue, Black or Grey ABS and sits on 4 self adhesive rubber feet. Incorporating guides for holding 1.5mm thick pcb, the base also has stand-off bosses for supporting small sub-assemblies etc. and ventilation slots. Front panel is held by 4 screws which run into integral brass bushes.

BIM6005 (143x105x55.5[31.5] mm) £2.32*
BIM6006 (143x170x55.5[31.5] mm) £3.08*
BIM6007 (214x170x82[31.5] mm) £4.12*

MULTI-PURPOSE BIMBOXES

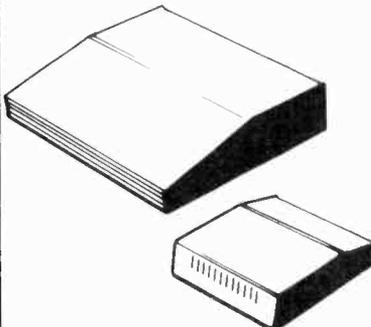
Moulded in Orange, Blue, Black or Grey
ABS with 1mm thick Grey aluminium
recessed front cover which is retained by
4 screws running into integral brass bushes.
1.5mm pcb guides are incorporated on all
sides and as with all ABS boxes they are
85°C rated. 4 self adhesive rubber feet
also included.

BIM 4003 (85x56x28.5mm) £1.24*
BIM 4004 (111x71x41.5mm) £1.56*
BIM 4005 (161x96x52.5mm) £2.08*



All aluminium, 2 piece desk consoles with either 15° or 30° sloping fronts, sit on 4 self-adhesive non slip rubber feet. Ventilation slots in base and rear panels permit efficient cooling.

Colour Code	Top Panel	Base
A	Off White	Blue
B	Sand	Green
C	Satin Black	Gold



15° Sloping Panel

BIM7151 (102x140x51[28] mm)	£ 9.43*
BIM7152 (165x140x51[28] mm)	£10.43*
BIM7153 (165x216x51[28] mm)	£11.42*
BIM7154 (165x211x76[33] mm)	£12.39*
BIM7155 (254x211x76[33] mm)	£13.66*
BIM7156 (254x287x76[33] mm)	£14.65*
BIM7157 (356x211x76[33] mm)	£15.80*
BIM7158 (356x287x76[33] mm)	£16.78*

30° Sloping Panel

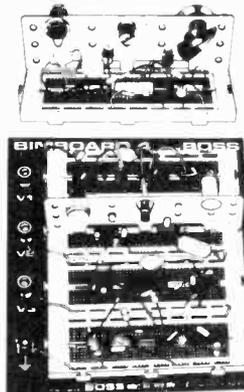
BIM7301 (102x140x76[28] mm)	£ 9.43*
BIM7302 (165x140x76[28] mm)	£10.43*
BIM7303 (165x183x102[28] mm)	£11.42*
BIM7304 (254x140x76[28] mm)	£12.39*
BIM7305 (254x183x102[28] mm)	£13.66*
BIM7306 (254x259x102[28] mm)	£14.65*
BIM7307 (356x183x102[28] mm)	£15.80*
BIM7308 (356x259x102[28] mm)	£16.78*

DIL COMPATIBLE BIMBOARDS

Bimboards accept all sizes of DIL packages as well as resistors, diodes, capacitors and LED's etc. They have integral Bus Strips running up each side for carrying Vcc and ground as well as Component Support Brackets for holding lamps, fuses and switches etc. Available as either single or multiple units, the latter mounted on 1.5mm thick, matt black aluminium back plates which stand on non slip rubber feet and have 4 screw terminals for incoming power.

Bimboard 1 contains 500 individual sockets whereas the multiple units containing 2, 3 or 4 Bimboards incorporate 1,100, 1,650 or 2,200 individual sockets, all arranged on a 2.5mm(0.1") matrix.

Bimboard 1 £ 9.72* Bimboard 2 £22.68*
Bimboard 3 £32.40* Bimboard 4 £42.12*



BIMDICATORS

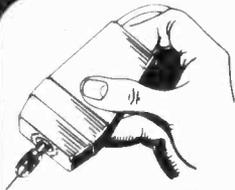


Remember we are also one of Europe's largest manufacturers of Filament, Neon and LED indicators. Send for our BIMDIATOR DATA

MAINS BIMDRILL

Operates directly from 220-240Vac and supplied with 2 metres long cable fitted with 2 pin DIN plug. Will drill brass, steel and aluminium as well as pcb's etc. Has integral biased-off switch and accepts tools with 1,2 and 3.2mm dia shanks £9.72*

Accessory Kit including 1mm, 2mm, .125" twist drills, 5 burrs and 2.4mm collet £2.20*



12 VOLT BIMDRILLS

2 small but powerful 12V dc drills, easily held in hand or used with lathe/stand adaptor. Both drills have integral on/off switches and 1 metre long cable.

Mini Bimdrill with 2 collets up to 2.4mm capacity £7.56*

Major Bimdrill with 3 collets up to 3mm capacity £12.96*

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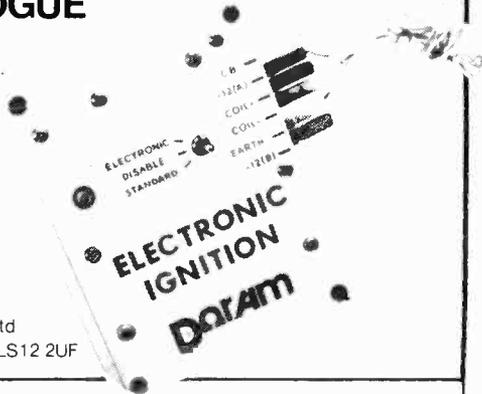
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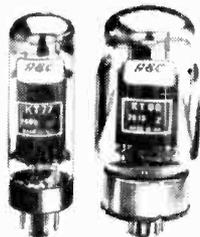
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Illustration actual size

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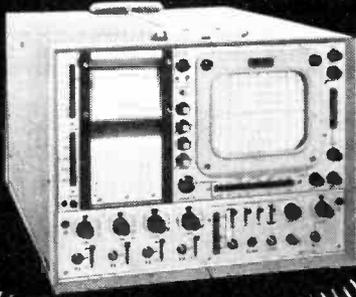
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A NEW INSTRUMENT FOR WIDENING INDUSTRIAL APPLICATIONS



The FOR-4 Mark 2

The new Medelec FOR-4-2 fibre optic recording oscilloscope is the result of a constant research and development policy. It incorporates many refinements which have been made to customers' special requirements.

The FOR-4-2 provides industrial and research users with high quality recording facilities at really low cost. X-Y Plot, Transient and Raster mode are all available in a single instrument.

Special features of the Medelec FOR-4-2 include:

- 10 times gain X and Y (1mV/cm on 4 Y channels)
- Fully automatic triggering (with higher sensitivity)
- Improved recording facilities (for greater flexibility)
- Light control filter (for excellent contrast)
- Wide speed range (from 0.1 to 1000 mm/sec—in 3 models)
- Internal loudspeaker (for audio monitoring)

For further information on the new FOR-4-2 or instruments in the range, contact:

MEDELEC LIMITED
Manor Way, Woking
Tel: Woking (048 62) 70331
Telegrams: Medelec, Woking

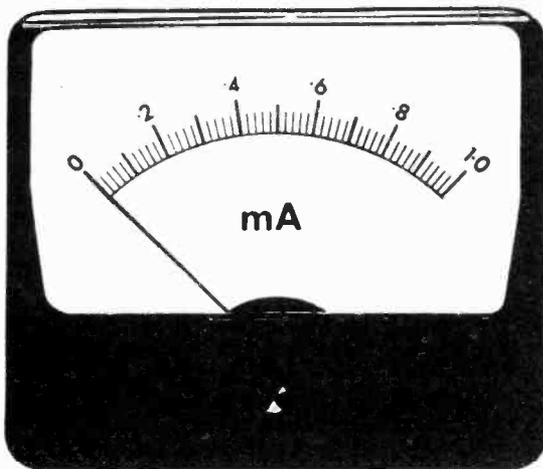
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Leaders in Fibre Optic Recording

WW—073 FOR FURTHER DETAILS

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LR70/71 bulk tape erasers are simple to operate and will erase cassettes, cartridges and reels of tape up to a maximum reel size of 11½" and tape width of 1", quickly and efficiently.

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— for the price of a signal generator!

If you're servicing or manufacturing mobile communications receivers then look closely at Farnell's new synthesized 10 to 520MHz signal generator, the SSG520. It outperforms other units in many respects, and, with its remarkable ease of setting, repeatability and SINAD facility gives you most of the advantages of a receiver test set costing much more.

Typical tests the SSG520 does with speed and simplicity are:—
 checking and aligning channel frequency and bandwidth
 i.f. and filter alignment
 sensitivity tests (so easy with SINAD)
 mute/squelch performance
 adjacent channel rejection (using two units) and
 signal to noise and a.f. distortion tests are made easier using the SSG520.

R.F. leakage is remarkably low, permitting totally unambiguous sensitivity measurements down to $0.2\mu\text{V}$ ($0.05\mu\text{V}$ if you like, using a 20dB pad). There's no leakage from counter display holes—there isn't one; you don't need a counter with our thumbwheel setting/readout. The SSG520 can be tuned in 100Hz steps under locked conditions with maximum stability over the entire frequency range.

Contrast this with competitive instruments which have either a mere 2% tuning range before re-lock or will only fast synthesize in 100kHz steps! Stability and accuracy are excellent and an optional ovened crystal version is available. Sideband phase noise is better than -100dB/Hz and harmonics better than -25dB . Any combination of a.m. and f.m. modulation, internal or external is possible. Output is calibrated and automatically levelled over the whole frequency range and the attenuator is set by adjacent 10 and 1dB click stop controls giving direct reading of dBm and volts—quicker to operate and enabling accurate mute/squelch settings.



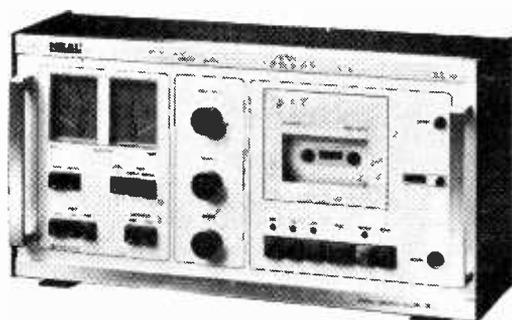
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So if you want to improve service or production throughput of mobile radio telephones, fixed receivers, handportables, public correspondence radiophone and paging systems without spending as much as you would have thought, use this magazine's reply system now to obtain full details and price.

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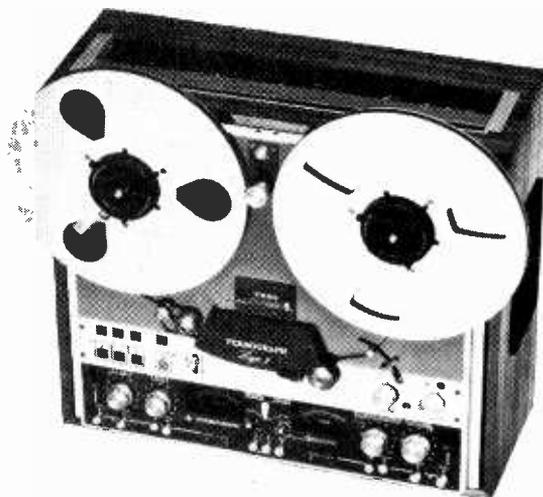
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With Neal Ferrograph you get the right equipment for the job, and the best in its class. A good formula for success, of which you can be assured every time you choose from the fully - integrated range of specialist recording and ancillary equipment in the NEAL FERROGRAPH range.



The NEAL 302.

Incorporating a 3-motor mechanism, controlled by a full solid state logic system actuated by ultra light touch buttons, this is the machine used by top recording studios and broadcasting stations, for quality cassette copies and for in - cassette duplication masters.



The Ferrograph Logic 7.

A transportable tape recorder of unrivalled facilities; taking all spool sizes up to 27 cm, and providing three speeds, plus positive action push buttons in association with logic circuits ... for fast, safe tape handling under all conditions.

Studio 8

A professional studio tape recorder logic controlled for superb tape handling characteristics, offering a choice of stereo, twin track and full or half track mono heads, PPM or VU meters, IEC (CCIR) or NAB equalisation, console or transportable models.



The RTS 2.

Combines in one easy to use compact instrument the measurement of gain, noise, frequency response, input sensitivity, output power, distortion and the parameters relating to recording equipment, such as wow and flutter, crosstalk, drift and erasure. Its range of application can be extended even further by the addition of the Auxiliary Test Unit ATU 1.

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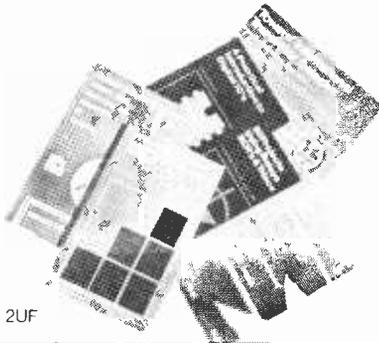


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Manufactured and guaranteed by Catronics Ltd

Model DFM 500, £164.50 + £5 carr + 8% VAT

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HIGH STABILITY OVENED VERSION. DFM 500/5 also available at £204.50. Our popular 200MHz Counter still available at only £137.50

REMOTE CONTROL FOR WW TELETEXT DECODER

A modification kit is now available to give Remote Control selection of all functions except setting the time select thumbwheel switches. The Remote Control Unit is housed in a cabinet approx. size 6" x 3" x 2" and connected to the main decoder by approx. 5 yds. screened cable.

PRICE ONLY £14.80 + 40p P&P

'Board 3' is also available as an additional unit to update the 'Wireless World' Teletext Decoder to give double height characters: colour background, conceal/ravel, etc., as described in December and January issue of 'Wireless World'

Our Kit includes plated-through hole P.C.B., all components and installation instructions. Price £33.68 + VAT (£3.47) + P&P (30p) = £37.45 total. P.C.B. available separately at £19.30.

Our main kits contain all the printed circuit boards and components necessary to build the complete decoder

A reprint of the series of articles is available at £1 50 + large 15p SAE (included free in complete kit)

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Set of 5 PCBs	£23.40	£23.35	30p
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COMPONENTS ALSO AVAILABLE SEPARATELY — SAE for price list

READY BUILT & TESTED DECODERS — £241.87 + £5 Carr

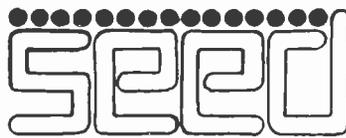
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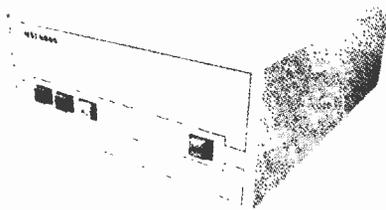
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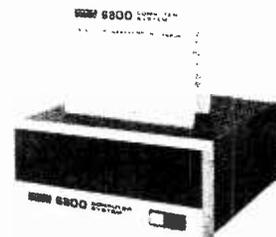
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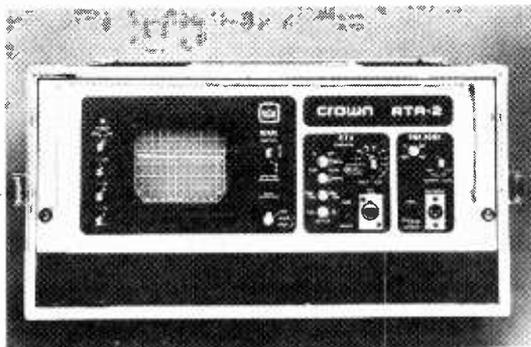
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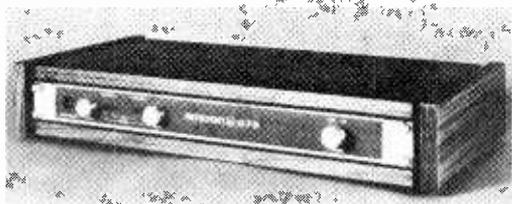
Real Time Analyser RTA2



The Amcron RTA2 Real Time Analyser is designed as much for use as a production tool as it is for on-site audio analysis of theatres, and recording studios. A flight case is available.

- ★ 5" CRT Display
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- ★ Price £1,960 ex. VAT

POWER AMPLIFIER D75



The AMCRON D75 power amplifier replaces the previous model D60. Employing completely new type circuitry it offers also many new features but without any increase in the price.

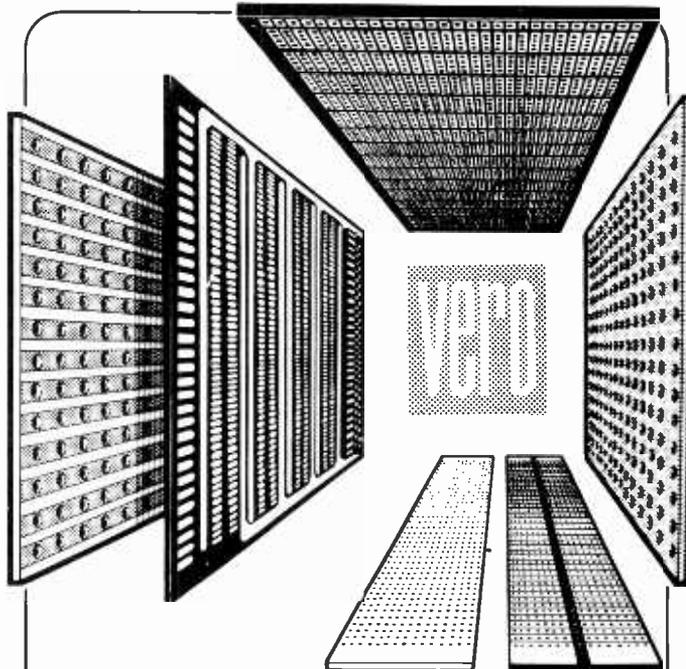
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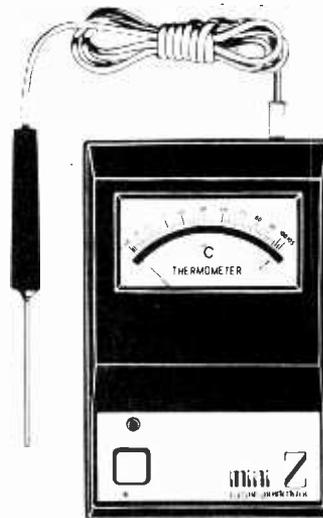


Our new 1978 catalogue lists circuit boards for all your projects, from good old Veroboard through to specialised boards for ICs. And we've got accessories, module systems, cases and boxes — everything you need to give your equipment the quality you demand. Send 25p to cover post and packing, and the catalogue's yours.

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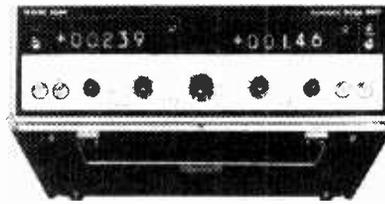
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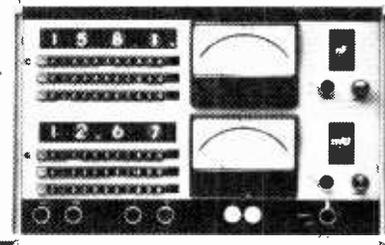
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Automatic Digital Bridge 1kHz 0.1%



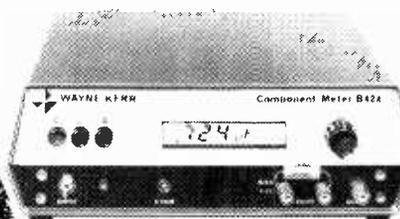
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- 1 2N5179 Transistor
- 2 UG-88/U BNC Connectors
- 1 PC Board

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- 2 UG-88/U BNC Connectors
- 1 LM/MC7805 Voltage Regulator
- 1 50volt 1Amp Bridge
- 1 LED Indicator
- 1 PC Board

And all other parts for assembly. **Now Only \$29.95**

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Crystal Filters. Tyco 001-19880 same as 2194F

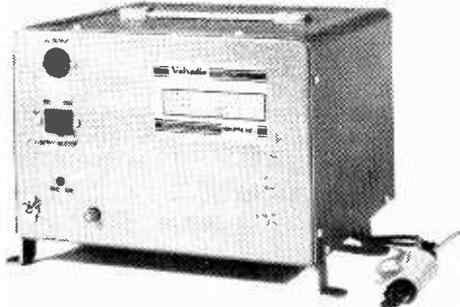
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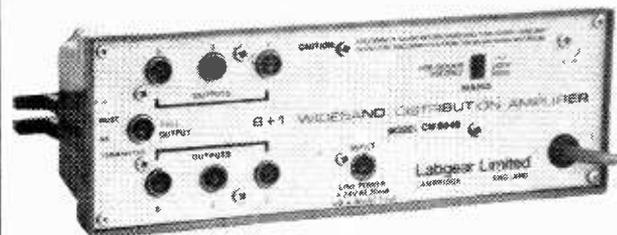
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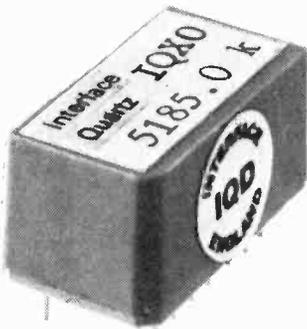


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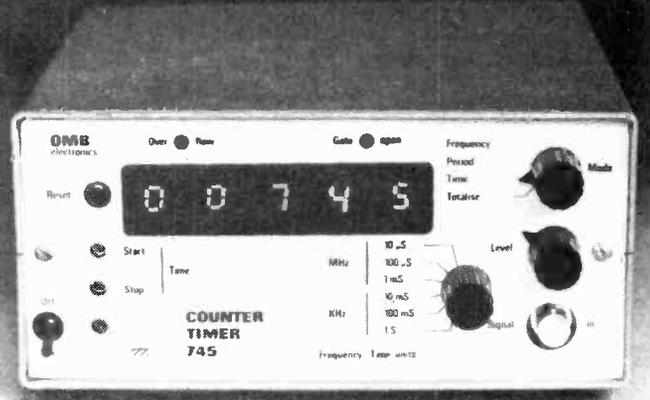
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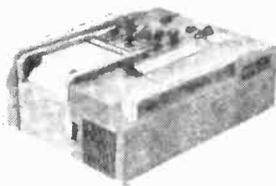
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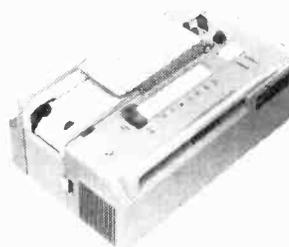


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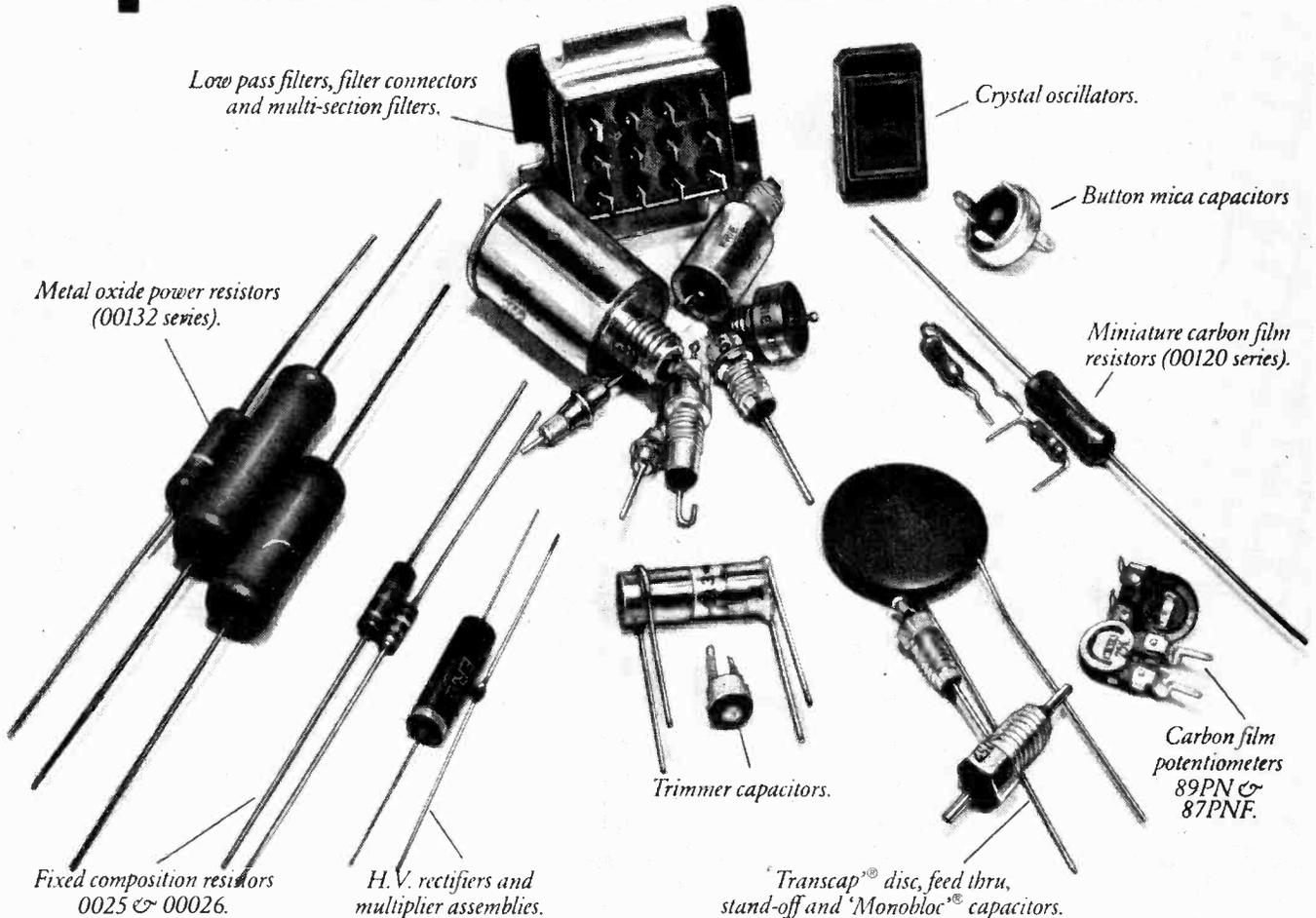
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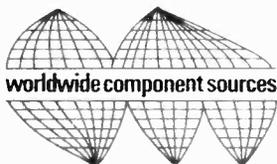


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IT IS UNLIKELY that many engineers have hitherto seen themselves as technological Renoirs or Gainsboroughs, or even prophets. But painters and practitioners in engineering suffer from the same barriers to full expression — they are often dependent on patronage and, in common with prophets, though most of them are not without honour, it is often not recognized in their own country.

We have recently seen examples of engineers who have originated quite remarkable inventions, but who, not having enough capital to put their ideas into practice, have hawked their wares round all the more obvious sources of finance — government establishments, manufacturers, financiers — with no success at all. No one, it appears, is in the risk business — at least, in the UK. As a result of this frustration, people are beginning to look overseas for their backing, which is fine for the inventor and his backer — not so good for this country.

Several British organizations exist for this very purpose, although if our correspondence pages are anything to go by, many bright ideas go unrecognized. A common complaint among engineers who do manage to sell their ideas and are assisted by, for example, the NRDC, is that the amount of money advanced is insufficient for an efficient operation. It may be said that half the amount needed is better than nothing, but if a cramped financial position leads to excessive caution and inhibits the broad view, it could well be worse than nothing.

Reasons for the directors of companies not wishing to risk venture capital on inventions with which they

are unfamiliar hinge to a large extent on the very fact of their unfamiliarity. Company directors, as a class, are not noted for their engineering knowledge, being recruited in the main from accountants, economists and arts graduates. Their field of interest is in marketing, finance and sales; the products of the companies over which they preside need not have much influence on their work at all, except insofar as they determine the people they rub shoulders with in business.

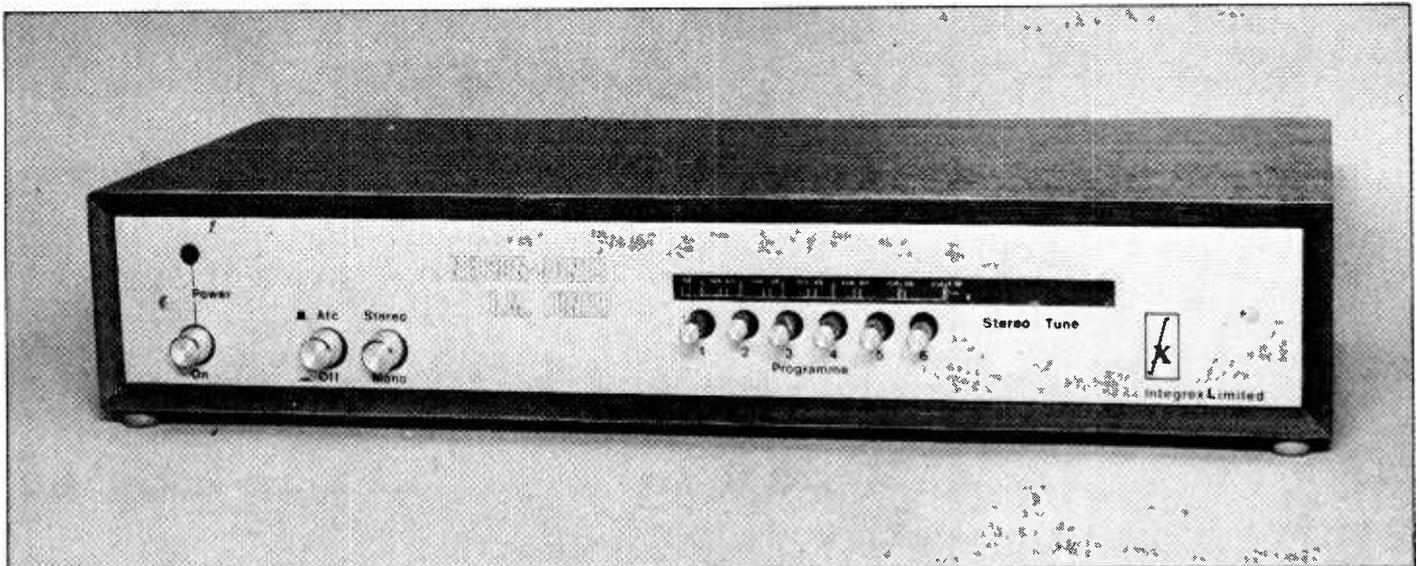
While they are reasonably adept in their own sphere of activity it seems unlikely that a financier is best able to judge the worth of even a simple piece of engineering, and if the project put forward for evaluation is even moderately recondite, then a degree of relevant knowledge is essential. And yet only around 30% of UK directors have any such knowledge. In contrast with this, Germany has about 70% of knowledgeable directors and the US 85%. Taken in conjunction with our fairly dismal performance in recent years, these figures are significant, although it is impossible to say that this is the real reason.

It does appear, though, that there is a need for more university and college students to be given a chance to read their chosen engineering subjects, with a background support of 'business' training. Admittedly, this does fly directly in the face of the university tradition which insists that universities are not there to train people, but to educate them. Times are hard, however, and it should be recognized that some students, at least, are going to have to soil their hands and engage in vulgar commerce.

Stereo f.m. tuner — Mk II

Improved design uses new i.cs — 1

by L. Nelson-Jones, F.I.E.R.E.



This tuner is based on the author's highly successful design first published in the April, 1971 issue. The circuit has been modified to use two recently introduced and improved i.cs, together with a pre-aligned f.e.t. front-end module designed by the author. The circuit features an improved a.f.c. system which operates directly on the varicap tuning and allows simultaneous tuning of all the r.f. stages. Circuit options include the use of either a six-pole LC filter or ceramic resonator i.f. units.

WITH THE LARGE SIGNAL LEVELS that can occur from high-gain aerial arrays it is possible for an f.e.t. front-end to malfunction. This is due to oscillator-pulling by the signal at g_1 of the mixer which has capacitive coupling to g_2 and therefore the oscillator. In extreme cases the oscillator may be pulled completely into lock with the signal, which results in a zero-frequency i.f. To overcome this a diode limiter has been used to damp the second tuned circuit so that the oscillator cannot be pulled-in. The diode limiter does, however, lower the image-frequency rejection performance but this only occurs at very high signal levels. With this modification, and by applying around 20dB of a.g.c. to the r.f. stage at high signal levels, the front end can handle signals of around 600mV.

This front-end circuit, shown in Fig. 1, also differs from the Mk I design because it does not have a separate a.f.c.

system. The tuning supply voltage is now modulated by the a.f.c. voltage, and controls all three tuned circuits together. Even with a high level of a.f.c., there will be no loss or gain within the holding range.

Choke coupling is used in the r.f. stage so that all tuned circuits are at d.c. ground potential. Because Mk II design is for varicap tuning only, a more compact layout has been possible. Decoupling has been improved by shorter lead lengths and a reduction in value of the decoupling capacitors to 470pF. This ensures that the capacitors do not come near to resonance. The complete front-end is now housed in a screened case to reduce oscillator radiation and pick-up from sources such as the i.f. strip, stereo decoder, and the demodulator.

An isolated coupling loop on the oscillator coil is brought out via two terminals on the main i.f. board. The loop provides a level of around 50mV and is designed to feed a digital counter at 50 Ω impedance. This level will not cause excessive oscillator radiation, and can be interfaced to a counter via a buffer stage. The second gate of the r.f. stage is brought out through a decoupling RC network for a.g.c. A suitable biasing network is provided if a.g.c. is not required (a.g.c. ref).

I.f. amplifier

The first i.f. amplifier stage is in the front-end already described, and pro-

vides a broadly tuned output at 330 Ω impedance. Fig. 2 shows the main i.f. circuit. The block filter can be the Toko six-pole LC filter, as shown in Fig. 3(a). In this case the additional series resistor R_2 has to be used to raise the source impedance to 1k Ω , and C_2 is placed at the filter input to provide the design source capacitance. Correct loading is provided by the input biasing resistor R_8 .

A second option, shown in Fig. 3(b), is to use a pair of Toko i.f. ceramic resonators, type CFSE-10.7, which have a design source impedance of 330 Ω . In this case R_2 and C_2 are not needed. As there is no d.c. path through these filters, C_4 is also redundant and is replaced with a link. Because the gain with these resonators is too high, a 10dB attenuator is placed between the two filter sections (this does not impair the noise performance of the tuner).

A third choice is to use two Vernitron FM4 i.f. ceramic resonators as in the Mk I design. These filters cannot normally be directly cascaded, but the 10dB attenuator section between them provides a satisfactory performance. Whichever type of ceramic resonator is used, it is essential that both have the same colour coding.

The main i.f. gain is supplied by the multi-stage limiting amplifier contained within IC₁. The circuit has been designed around the recently introduced CA3189E, which is an improved and

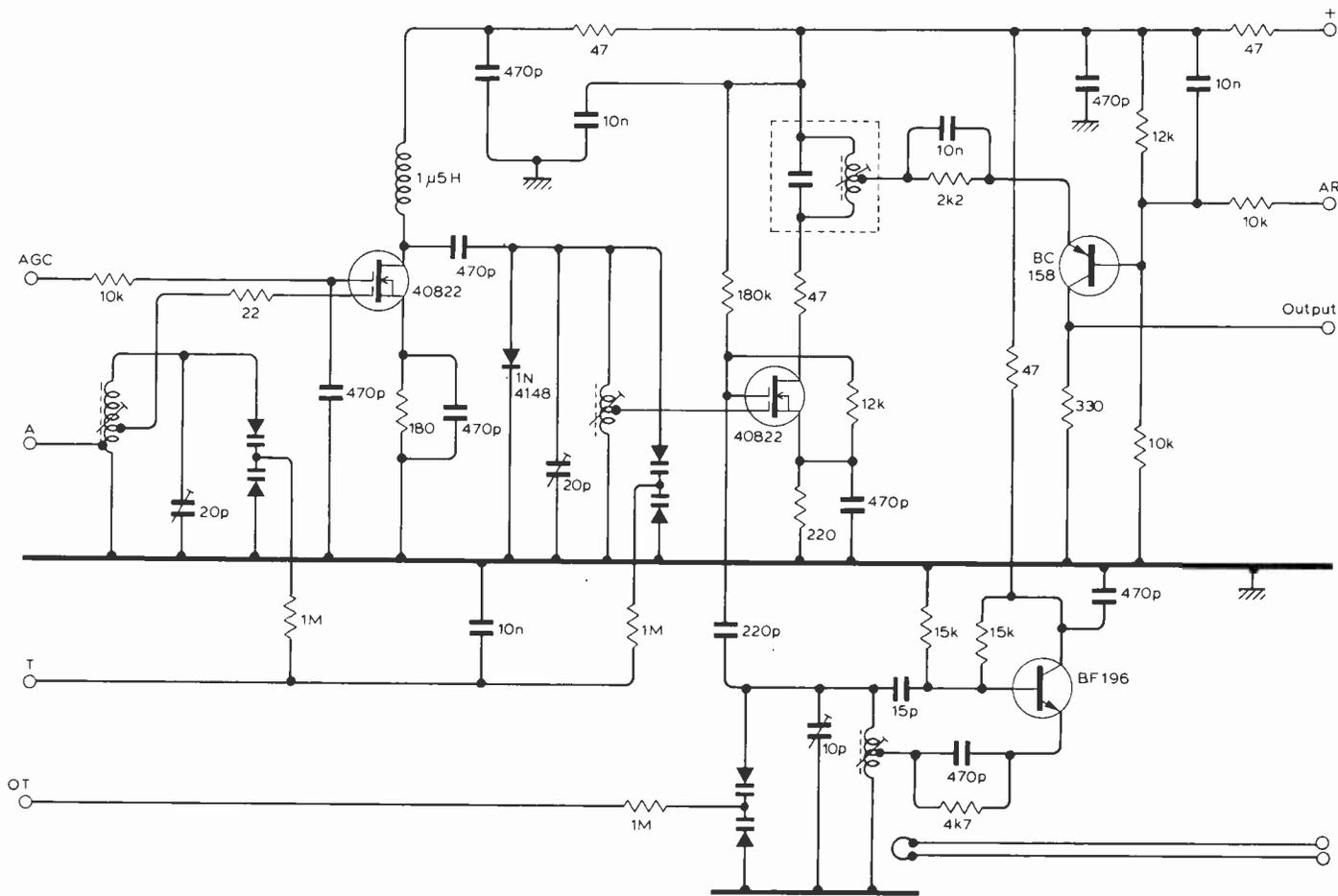


Fig.1. Front-end module. Damping the second tuned circuit with a diode improves signal handling capacity.

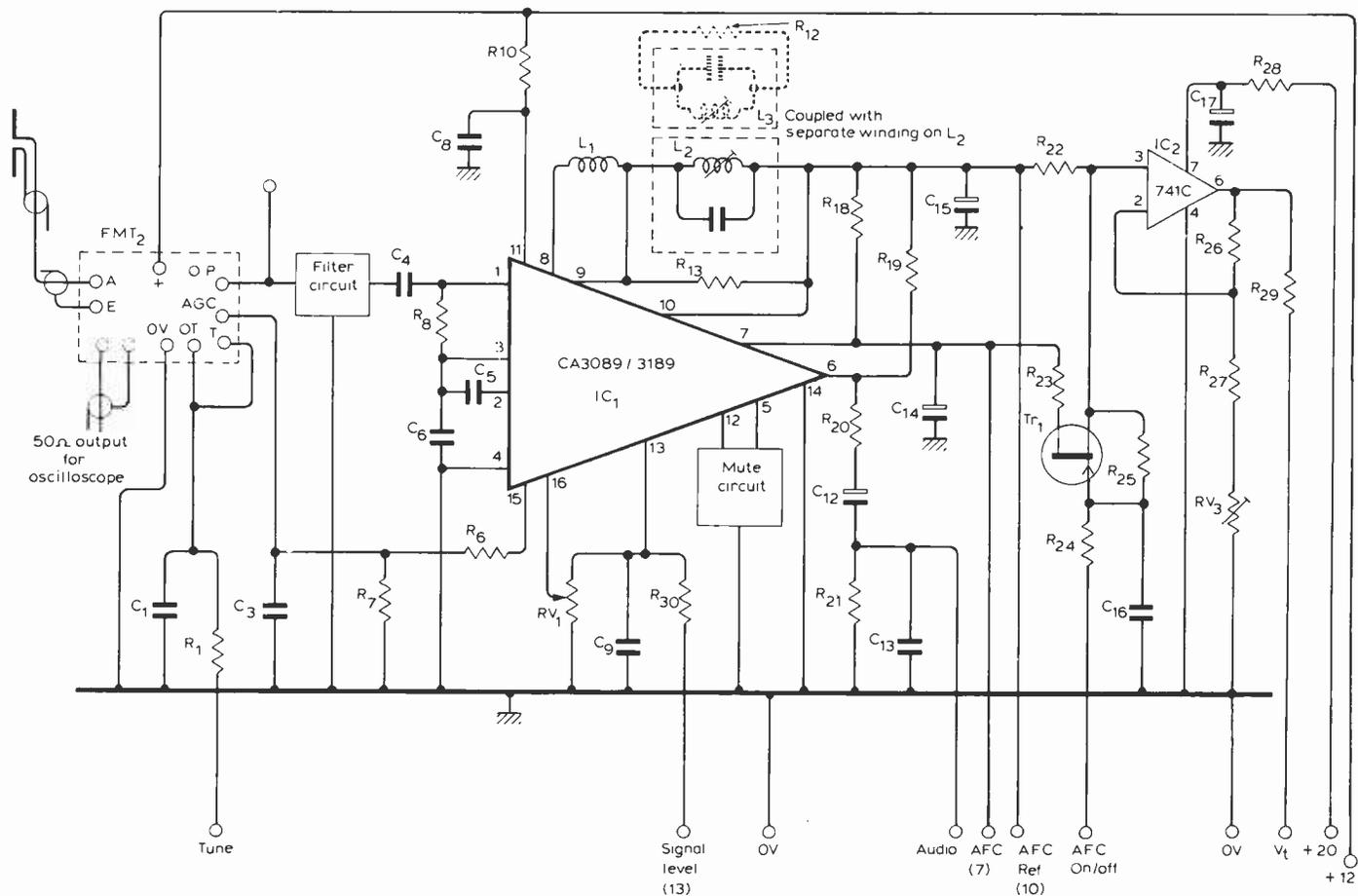


Fig.2. Main i.f. circuit can use either the CA3089E or the newer CA3189E.

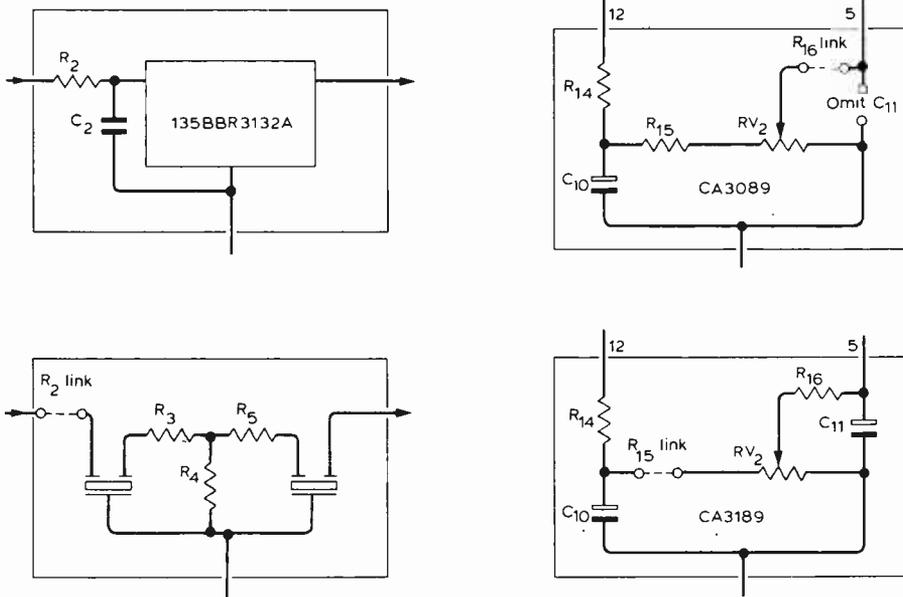


Fig.3. (a) Six-pole LC filter. (b) Two ceramic resonators. (c) Muting circuit for CA3089 and (d) CA3189.

Fig.4. Stereo decoder. Emitter follower on pin 11 of the i.c. improves the signal-to-noise ratio at low signal levels by reducing the separation.

somewhat altered version of the CA3089E. The printed circuit board can be used with either of these i.c.s by making suitable component changes. However, in my experience the performance of the CA3089E is rather inferior to the CA3189E. The audio output of the CA3189E is adjustable and the values used give a level of 490mV for each of the options.

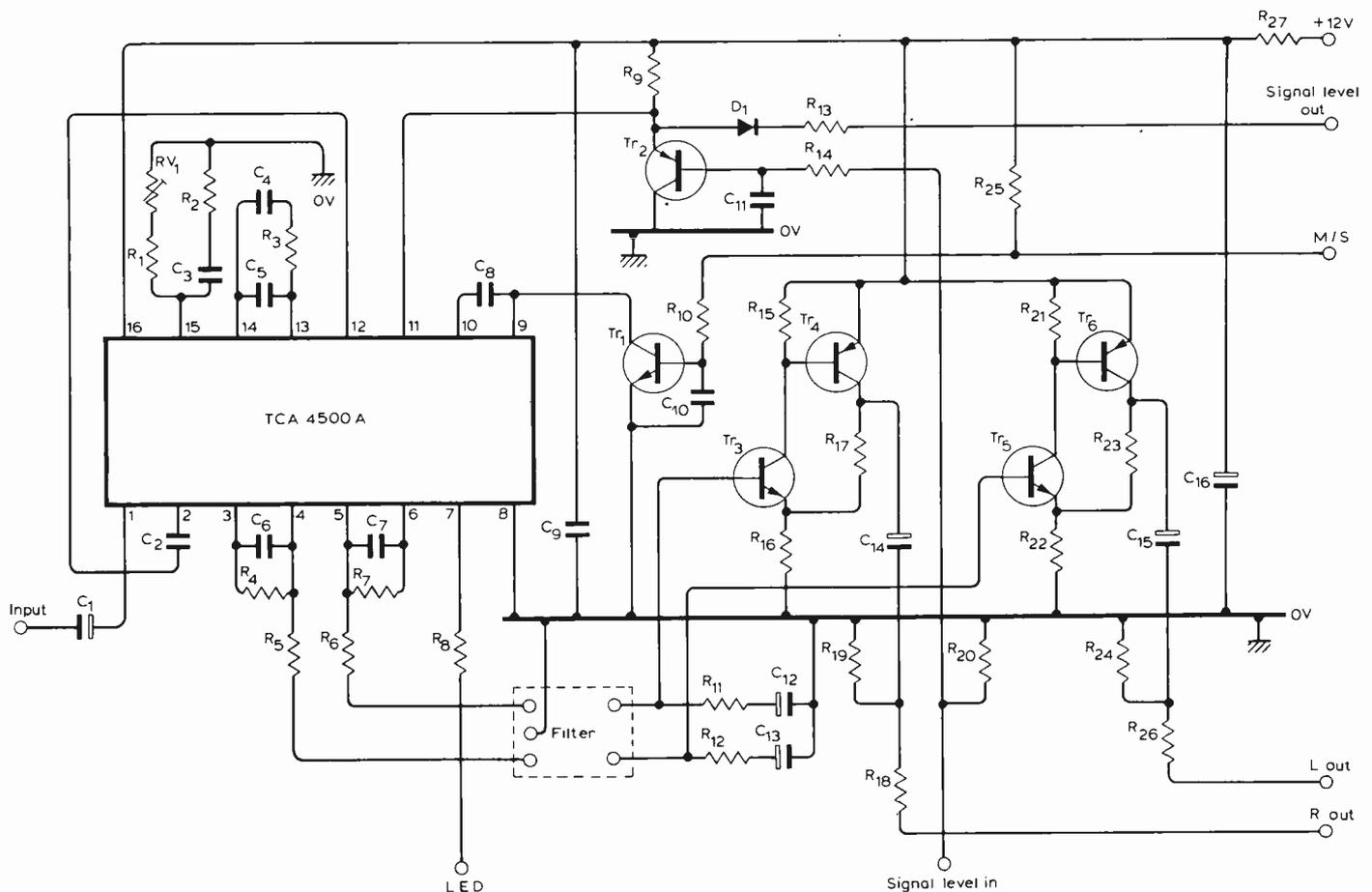
A further option is available for the CA3089E and CA3189E because both can be used with single or double-tuned quadrature coils, L_2 and L_3 . The double-tuned arrangement can give very low demodulation distortion if correctly

adjusted, but this requires a low distortion f.m. signal generator, and distortion measuring equipment. With a single-tuned circuit both i.c.s will still give low distortion compared to earlier integrated circuits such as the TAA661B. Both devices provide an a.g.c. feed for the front-end module, but because the level is not the same for the two devices the a.g.c. voltage is fed through potentiometer R_6 , R_7 . The a.g.c. threshold is adjustable with the CA3189E. I have found this particularly useful in setting up the signal strength output, from pin 13 of IC₁, to give a steady and progressive increase of level with signal input voltage. The potentiometer concerned is usually set around midway.

The external muting circuit components for the two i.c.s differ and are shown in Fig. 3(c) and 3(d). The newer CA3189E operates on deviation as well as signal level, and the value of R_{18} sets the deviation at which muting begins. This may be varied if required from about 2.7k Ω for a large deviation before muting, to 22k Ω for a very small-deviation before muting. In both circuits the setting of RV_2 determines the signal level at which muting takes place. With RV_2 set to 0V, the muting action is stopped.

Tuning voltage supply and a.f.c.

The a.f.c. output from IC₁ is not applied directly to the front-end, but is used together with the a.f.c. reference output from IC₁ to derive a 12V tuning supply which is modulated by the a.f.c. output



of IC₁. It should be noted that the tuning range of the new front-end is 1.5 to 11.5V for 87.5 to 108MHz. The a.f.c. reference of the CA3089E and CA3189E is a 5.6V zener-stabilized supply which can be used for the reference of a conventional stabilized supply. If a suitable amount of a.f.c. voltage from pin 7 of IC₁ is added to the reference supply, then the tuning voltage will change in a way that will correct the error which caused the a.f.c. output to differ from the a.f.c. reference voltage level.

The a.f.c. reference is permanently connected to the non-inverting input and the op-amp IC₂ via an f.e.t. so that it can be switched in and out. When the f.e.t. is switched on, R₂₂ and R₂₃ control the amount of a.f.c. voltage that is added to the reference level. The tuning voltage output is controlled by the feedback chain R₂₆, R₂₇, RV₃. The a.f.c. switch operates with the source and drain of the f.e.t. at the a.f.c. reference level. If the gate is connected to 0V the f.e.t. is biased off; if the gate is left free, R₂₅ turns the f.e.t. on; an f.e.t. with a cut-off bias of less than -4V has been chosen to ensure clean switching. Values chosen for R₂₄ and R₂₅ allow the a.f.c. switch to be formed by a pair of

touch-pad contacts between 0V and the a.f.c. on/off input. Capacitor C₁₆ ensures that this input is not excessively sensitive to interference. If touch-pad operation is used it is essential that the chassis is correctly earthed and connected to the 0V line.

With this system an almost constant a.f.c. performance is achieved across the whole f.m. band because, as the tuning voltage is reduced, the amount of change due to any a.f.c. control is also reduced. The tuning characteristic of the front-end is almost perfectly logarithmic with tuning voltage. The tuning-voltage supply must be free of noise and hum to prevent spurious modulation of the oscillator frequency. This is achieved by the high common-mode supply rejection of IC₂, and by the filter R₂₈, C₁₇ in the supply to IC₂.

Stereo decoder

The decoder circuit in Fig. 4 is essentially the same as that published in the April 1978 edition by M. J. Gay. The capacitor between pins 2 and 12 has been made 10nF rather than 6.2nF quoted, and the input capacitor has been raised from 2μF to 4.7μF. These changes were made because my stock did not contain

the original values, so they need not be implemented.

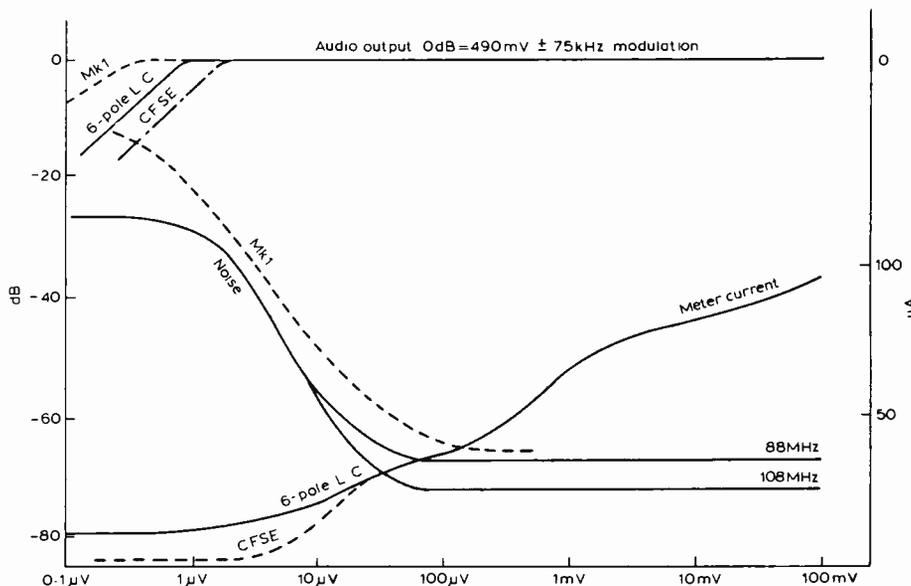
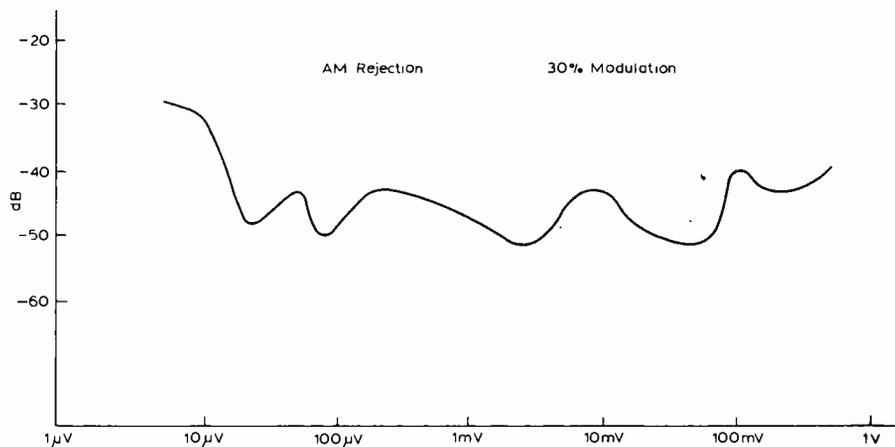
The TCA4500A has a low impedance output due to the feedback networks R₄, C₆ and R₇, C₇. It is therefore necessary to feed the multiplex filter through resistors which have a value equal to the design source impedance of 4.7kΩ for the filter. Because the design load impedance of this filter is also 4.7kΩ, there will be at least a 6dB loss. To restore the audio output level, and to isolate the filter from the load, an amplifier with a gain of two is connected to each of the stereo outputs. High negative feedback in each of these amplifiers makes the input impedance very high, so the matching network R₁₁, C₁₂ and R₁₂, C₁₃ is connected to ensure that the filter sees a resistive termination of 4.7kΩ. The bias for these two stages is derived from the d.c. output level of the i.c. at pins 4 and 5. The filter has a low d.c. resistance from input to output and the values of R₅ and R₆ are insufficient to cause any appreciable loss of voltage to the bases of Tr₃ and Tr₅.

Gain of the stages is defined by the equal values of load resistors R₁₆, R₁₇ and R₂₂, R₂₃ so only half of the output is

Receiver performance

Frequency range	87.5 to 108MHz (tuning voltage +1.5 to +11.5V)
I.f.	10.7MHz
I.f. bandwidth (-3dB)	220kHz (CFSE-10.7 filters-FM4)
	250kHz (6-pole LC filter Toko 135BBR3132A)
Input impedance	nominally 75 ohms unbalanced
Limiting input signal threshold	1μV typical
-30dB quieting level (mono)	1.5μV typical
Capture ratio	1dB
Image response	-48dB
I.f. response at input	about -100dB
Oscillator voltage at aerial input	less than 1mV
Oscillator output to counter	about 50mV into 50Ω
Muting threshold range	adjustable from 0 to about 8μV
Audio output level	490mV for ± 75kHz peak deviation
Spurious decoder outputs	better than -60dB at 19kHz and all harmonics
Audio frequency response	10Hz to 15kHz ± 1dB
De-emphasis time constant	50μs (75μs with capacitors raised by 50%)
A.f.c. pull-in range	± 500kHz (1mV input level)

Mark II tuner is better than the Mk I design on noise performance, especially at low signal levels, and this is very noticeable in listening tests. A.m. rejection of the new tuner is again better especially in listening tests.



fed back to the emitter of the input transistor. The output d.c. level is blocked by C_{14} and C_{15} , while R_{19} and R_{24} prevent clicks if the output is connected to an amplifier or switched after the receiver has been turned on. Resistors R_{18} and R_{26} in the output leads prevent oscillation in these two amplifier stages if very long and capacitive leads are used.

Stereo/mono switching is achieved by transistor Tr_1 . This is normally biased on, and the decoder is in the mono state. If the m/s input is grounded, the biasing is removed and Tr_1 is turned off, which restores the decoder to stereo operation.

Capacitor C_{10} , together with the biasing resistor R_{10} , form a h.f. filter to prevent this input from causing interference. A switching transistor was used because the manufacturers' data specifies a maximum capacitance, from pin 9 to ground, of 100pF and I felt that this was too easily exceeded. Also, a long lead into the 228kHz oscillator circuit is undesirable.

The signal-level voltage from the CA3189 can be used, as detailed in the TCA4500A article, to reduce the separation at low signal levels and provide a better signal-to-noise ratio. This has been added to the decoder circuit by placing an emitter follower between the signal level input from the CA3189, and pin 11 of the TCA4500A. Use of a p-n-p emitter follower ensures that the current can be drawn out of pin 11, and also provides a low impedance drive for the signal strength meter. The recommended resistor value for R_{13} is 39k Ω , which gives close to 100 μ A for full signal strength. This resistor may be reduced in value for movements up to 1mA. Diode D_1 is included to remove the V_{be} of Tr_2 from the feed to the meter. Resistor R_{14} and capacitor C_{11} are included to prevent interference from passing through Tr_2 . The supply to the decoder is filtered to prevent switching surges from reaching the rest of the tuner.

A stereo indication i.e.d. is driven from pin 7 through R_8 and a further 270 Ω on the tuning indicator board. If this board is not used, and only the external i.e.d. to the +12V supply is required, then R_8 should be raised to 680 Ω .

Tuning indicator

The circuit in Fig. 5 is exactly as used with the Mk I toner. It consists of a long-tailed pair feeding two i.e.d.s, whose brilliance will be equal for equal input voltage levels to the bases of the two transistors. A degree of degeneration is applied to lower the gain from the unbypassed resistors in the emitters. The output terminal is connected to the a.f.c. pin of the i.f. board. The stereo indicator i.e.d. in the original circuit had to be compatible with a filament lamp to match the Portus and Haywood decoder, but this requirement is not now

Parts List

Front end and i.f.

Resistors — 5% 1/4E carbon film unless otherwise stated

R 1	100k Ω
6	1k Ω (3089) 39k Ω (3189)
7	omit (3089) 39k Ω (3189)
8	560 Ω (6-pole) 330 Ω (CFSE/FM4)
10	47 Ω
12	2.7k Ω (double-tuned) omit (single-tuned)
13	18k Ω (double-tuned) 3.9k Ω (single-tuned)
18	4.7k Ω (3089) 8.2k Ω (3189)
19	omit (3089) 4.7k Ω (3189 double-tuned) 6.8k Ω (3189 single-tuned)
20	link (stereo) 4.7k Ω (mono)
21	100k Ω
22	5.6k Ω
23	33k Ω
24	2.2M Ω 10%
25	10M Ω 10%
26	10k Ω 2%
27	6.2k Ω 2%
28	680 Ω
29	47 Ω
30	1k Ω (stereo, or to suit meter on mono)

Capacitors — 20% tolerance unless otherwise stated

C 1	0.1 μ F	250V	polyester
3	10nF	100V	min. ceramic
4	10nF	100V (6-pole)	min. ceramic
	link	(CSFE or FM4)	
5,6	10nF	100V	min. ceramic
8	47nF	250V	polyester
9	10nF	100V	min. ceramic
12	10 μ F	25V	tantalum bead
13	33pF	100V (stereo)	min. ceramic
	4.7n	100V (mono)	10% ceramic
14,15	10 μ F	25V	tantalum bead
16	10nF	100V	min. ceramic
17	100 μ F	16V	min. vertical electrolytic

Other components

Tr_1	Silicon 2N4339 metal can n-channel f.e.t. or E113 plastic n-channel f.e.t.
RV_1	47k Ω min. horizontal skeleton potentiometer (3189 only)
RV_3	4.7k Ω min. cermet horizontal potentiometer
L_1	22 μ H min. choke Sigma Products type SC10-22 μ H
L_2	TKACS 34342BM } (double tuned for 3089 or 3189)
L_3	TKACS 34343AUO } Toko UK Ltd
L_2	KACSK586HM (single tuned for 3089 or 3189) — Toko UK Ltd
Front-end module	Key Electronics FMT2-0 — Integrex Ltd

I.f. filters

Fig.3(a)	
R_2	680 Ω
C_2	10pF 100V min. ceramic
Filter	Toko 135BBR3132A

Fig. 3(b)

R_2	Link
C_2	omit
R_3	220 Ω
R_4	150 Ω
R_5	220 Ω
Filters	Toko CFSE or Vernitron FM4 (identical colour coding)

Muting circuits

3089 Fig. 3(c)

R_{14}	470 Ω
R_{15}	120k Ω
R_{16}	link
RV_2	470k Ω min. horizontal skeleton
C_{10}	0.47 μ F 35V tantalum bead
C_{11}	omit

3189 Fig 3 (d)

R_{14}	470 Ω
R_{15}	link
R_{16}	47k Ω
RV_2	10k min. horizontal skeleton
C_{10}	47 μ F 6.3V tantalum bead
C_{11}	2.2 μ F 35V tantalum bead

Stereo decoder

Resistors — 5% 1/4W carbon film unless otherwise stated.

R 1	10k Ω	2%	1/2W metal oxide
2	100 Ω		
3	1k Ω		
4,7	5.1k Ω	2%	1/2W metal oxide
5,6	4.7k Ω	2%	1/2W metal oxide
8	270 Ω		
9	4.7k Ω		
10	33k Ω		
11,12	4.7k Ω	2%	1/2W metal oxide
13	39k Ω		(for 100 μ A meter)
14	4.7k Ω		
15	6.8k Ω		
16,17	2.2k Ω	2%	1/2W metal oxide
18	220 Ω		
19	100k Ω		
20	4.7k Ω (3089) omit (3189)		
21	6.8k Ω		
22,23	2.2k Ω	2%	1/2W metal oxide
24	100k Ω		
25	33k Ω		
26	220 Ω		
27	22 Ω		
RV_1	4.7k Ω		min. horizontal cermet skeleton potentiometer

Capacitors — 20% tolerance unless otherwise stated

C 1	4.7 μ F 35V	tantalum bead
2	10nF 250V	polyester
3	220pF 160V	5% polystyrene
4	0.47 μ F 250V	10% polyester
5	0.22 μ F 250V	polyester
6,7	10nF 250V	5% Siemens polyester or polycarbonate
8	0.22 μ F 250V	polyester
9	10nF 250V	polyester
10,11	10nF 100V	min. ceramic
12-15	22 μ F 16V	tantalum bead
16	220 μ F 16V	min. vertical electrolytic

Other components

IC	TCA4500A, Motorola
$Tr_{1,3,5}$	BC109, BC149, ZTX109, BC184, etc.
$Tr_{2,4,6}$	BC179, BC159, BC214 etc.
D_1	1N4148
Filter	Toko BLR3107N

Tuning indicator

R_1	4.7k Ω
2	100 Ω
3	330 Ω
4	100 Ω
5	1k Ω
6	omit
7	270 Ω
$Tr_{1,2}$	BC109, BC149, ZTX109, BC184 etc.
$D_{1,2}$	0.2in red i.e.d.
D_3	0.2in green i.e.d.

Power supply

Transformer	RS Components 6VA type 196-296 secondaries in series, or type 196-303 secondaries in parallel
Rectifiers	1N4003 (four)
Capacitors	0.1 μ F 250V polyester (three) 470 μ F and 1000 μ F 25V min. vertical electrolytic
Resistor	22 ohm
Regulator	type 7812 plastic i.c.
Fuse	20mm 250 or 300mA anti-surge

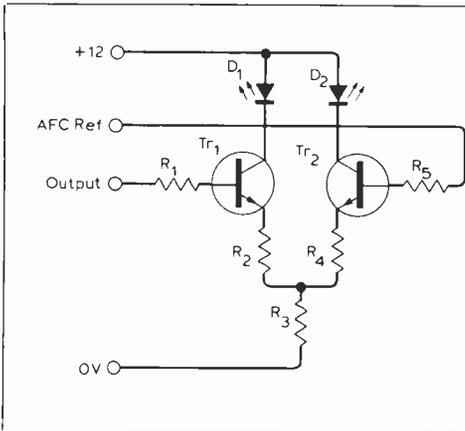


Fig.5. L.e.d. tuning indicator and stereo indicator.

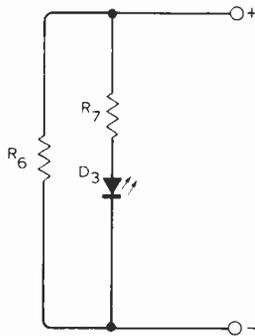


Fig.6. Power supply.

needed, therefore R₆ is omitted and R₇ is 270Ω.

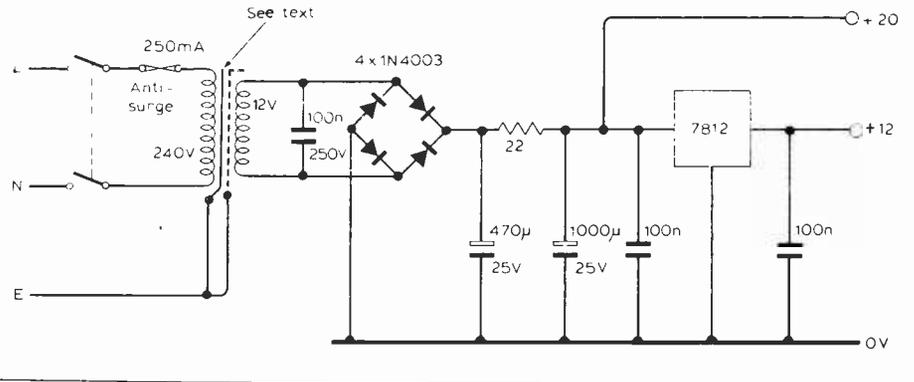
Power supply

The power supply shown in Fig. 6 uses a full-wave rectifier which feeds two smoothing capacitors. This arrangement produces very little ripple on the +20V supply to the a.f.c./tuning voltage circuit. The +12V supply to the tuner and decoder is stabilized by a 7812 i.c. which has two 0.1μF capacitors across its terminals for h.f. stability.

The capacitor wired across the bridge rectifier input prevents hole-storage noise in the diodes and forms a useful h.f. filter in conjunction with the leakage reactance of the mains transformer.

Interconnection of the tuner

The method of interconnection is shown in Fig. 7. It is important to ground the aerial input socket to the chassis only through its connection to the i.f. board. The 0V terminal and wiring must be grounded to chassis at



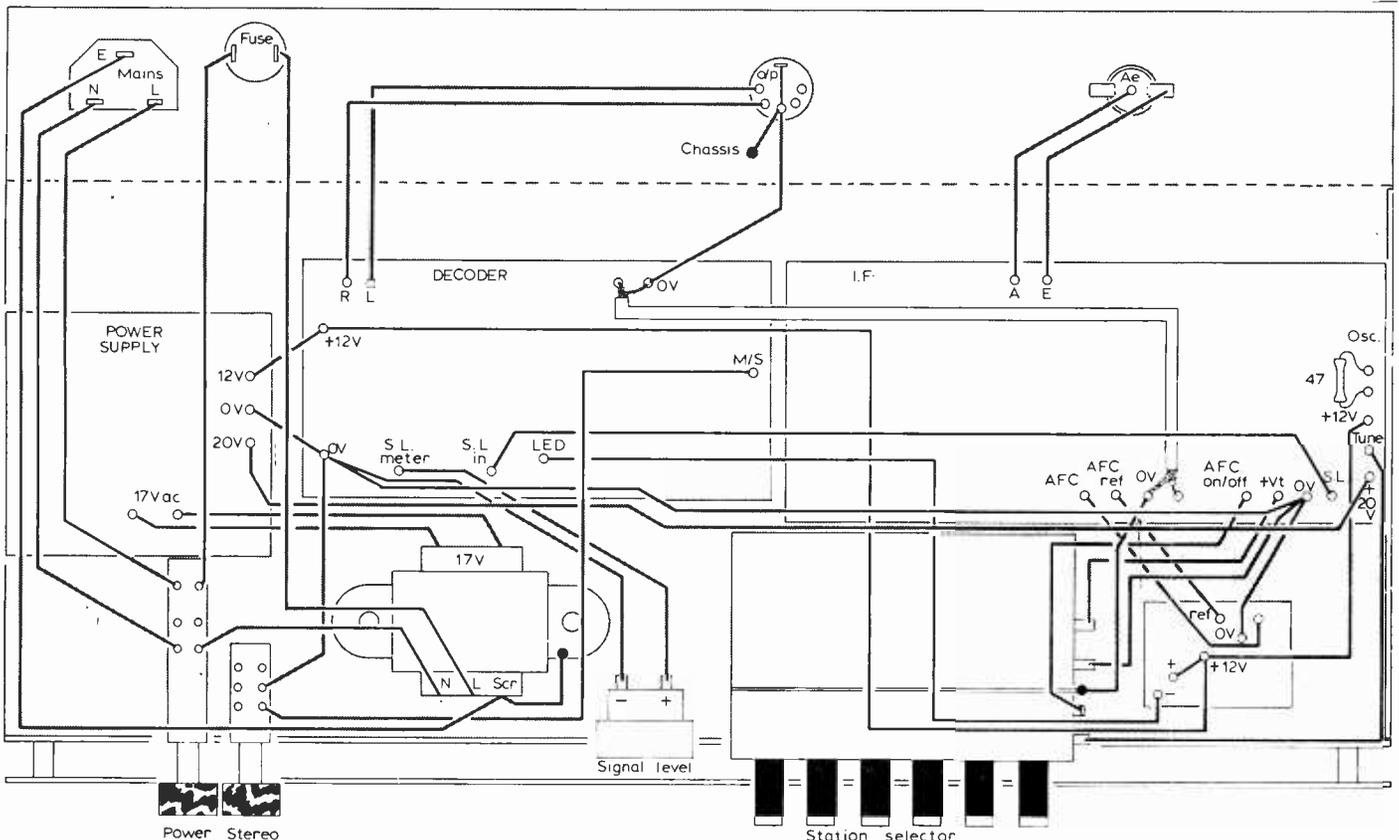
only one point near to the output. The mains earth must be connected only to the mains transformer frame and the interwinding screen. The transformer frame is insulated from the chassis because the tuner will normally be part of an audio system, and it is usual to connect only the pre-amplifier to the mains earth.

To make such an earthing system safe it is essential to use a transformer of

good construction which has a copper-foil interwinding screen, and a good earth bond to the frame. □

To be continued. The concluding article describes the alignment procedure and shows the printed circuit board layouts.

Fig.7. Interconnection diagram.



Instant tuning

By Cathode Ray

SEVERAL TIMES LATELY I have overheard on the radio hints of a great new idea: instead of changing from one radio* channel to another by turning a knob, meanwhile scrutinizing the movement of a pointer along a scale crowded with figures and names of places (quite likely obsolete) to find the position that will give the wanted programme, and then when at last it has been found, via a cacophony of intervening hisses and shrieks, finding it is too late to hear the start; instead, as I say, of this gruelling process several times every day, some inspired scientist (never any mention of an engineer, of course) is on the verge of making possible an instant switchover from one programme to another!

Well, of course, I thought this was just another example of how ill-informed so many people are, so that it can be commonly supposed that television was invented about 1950 and that Lindbergh was the first person to fly across the Atlantic (instead of, wasn't he, about the 55th?). But a few days ago I was pulled up short by reading in the IEE's journal *Electronics & Power*, July 1977 issue, p.547, an article by Mr James Redmond, Director of Engineering of the BBC (and therefore presumably knowing something about these matters) in which he said as follows:

'At one time there was a large audience of television viewers who remained tuned to one programme because they dared not retune the set in the hope of finding something more to their taste. Pushbutton television receivers quickly freed them from this

tyranny, and now we impatiently await the pushbutton portable v.h.f. radio.'

'Impatiently await!'

I fully appreciate the point about being afraid to depart from a channel, once it has been tuned in, because of the difficulty experienced by the lay public in retuning any radio (in its correct inclusive sense) receiver. My sympathies are entirely with them. What sort of a radio engineer is he/she who designs a set that obliges them to do anything so crude?

In the 9th February 1939 issue of *W.W.* (i.e., getting on for 40 years ago) I am on record as stating 'A certain trade list of receivers now on the British market describes 665 models. Of these, 231 are to be found with pushbutton tuning.' I went on to say 'As regards the date of invention, leaving out the inevitable Chinese and Egyptian claims to priority . . . it can definitely be said that it [pushbutton tuning] was on the market at least as early as 1928' (fifty years ago now) and went on to describe an American Zenith model so fitted, which enjoyed a ready sale, including one to me.

By 1940 (December issue of *W.W.*, p.499) I made fun of the 'ever-patient British public grinding away at their tuning controls' to change from one programme to another. Evidently the 231 pushbutton models less than two years earlier had not prevailed against the 434 others.

But Mr Redmond was writing about v.h.f. radios, which these early p-b models were not. So v.h.f. p-b models may still be in the future?

My wife being one of the lay public and therefore, I consider, fully justified in demanding instant programme changing, I provided her with it almost from the start; i.e., when I married her, which was not yesterday, since our grandchildren are by now doing a bit of demanding of their own in the electronic field. When the three BBC channels became available on v.h.f. (I exclude Radio 1 from consideration, while realizing that for many members of the population this is the preferred channel, but would I be hopelessly wrong in supposing that even they *sometimes* want a change?) I fully accepted the justice of the BBC claim that from then on all right-thinking citizens should turn exclusively to v.h.f. So, nearly 20 years ago, I scrapped all else and provided my wife with an all-v.h.f. set having a three-position switch covering her 'radio' needs.

All went happily until the BBC (obliged, I am sure, by hidden political

forces) began introducing intensely unwanted Open University lessons into v.h.f. channels. In distress at this unpredictable and frustrating development, my ever-loving wife turned to me; and with unfailing resourcefulness I devised a m.w. unit for Radio 4 making use of the existing switchgear.

This worked reasonably well until the next crisis, when for a time *all* Radio 4 channels, v.h.f. and m.w., were belting out studies on the life-style of the arachnidae or some such esoteric matter about which my wife did not at that time wish to know. Eventually it transpired that this was not more than a technical hitch lasting for a mere couple of hours, but it undermined my reputation for foreseeing and providing against every eventuality.

For the last year and more, the BBC, no doubt at the receiving end of bitter complaints from those listeners who had done as they had been bid and had changed over to v.h.f., has used every available gap between programmes to plug the necessity for a *three-waveband* receiver, long-wave, medium-wave and v.h.f., in order to be able to hear everything they provided. (I hope no unprincipled radio dealers take advantage of customers who remember only the 'three-waveband' bit, by unloading stock that technically conforms to this description but lacks v.h.f. and includes instead the not widely demanded short waves.) Among the ordinary indiscriminating British public these exhortations are unlikely to have been heeded, because the OUB public are able to get everything from the BBC on their cheap imported 'transistors,' many of which are *one-waveband* sets. Any whose interest may have been kindled to the extent of inquiring would quickly be deterred by the price of the a.m. plus f.m. models.

Some of the OUBP—housewives chiefly — got their first jolt when they suddenly found that their Tuesday and Thursday afternoon plays had turned into Questions in Parliament (cleverly disguised as live broadcasts from the Zoo) and 'Disgusted, Tunbridge Wells' became thick with complaints. The BBC could always say, in polite euphemistic terms, of course, 'We told you so.' The same thing will happen again in November, only much more so, among the one-waveband brigade when, not having taken in the oft-repeated warnings, they find themselves unable to get *any* Radio 4 programmes at all, unless they acquire new sets at least with long waves. Moreover, those of

*About 25 years ago I poured scorn on the illogicality and sheer ignorance of persons on the entertainment side of broadcasting who started a custom of distinguishing between programmes and receivers confined to sound and those that provided both sound and sight by calling the former *radio* and the latter *television* (or more usually TV). How did the poor mutts imagine that TV programmes were broadcast if not by radio? Were it not that they, and the non-technical British public who swiftly copied them, would have been unaware that this absurd nomenclature needed any excusing, they might have excused themselves by the plea that there was no word ready for use for referring to sound broadcasting. All except purists might have accepted 'telesound' had not 'tele' (pronounced 'telly') come to convey to the said public the idea, not, as it should, of distance, but of 'the box' and especially its pictures. Anyway, at last I have given up the unequal struggle, and can only gaze in awe at the respect for principle exhibited by the BBC in retaining the name *Radio Times* for their weekly list of all radio programmes, with or without pictures. Come to think of it, the same inflexible devotion to a cause is displayed by *Wireless World*. And by the author of *Foundations of Wireless and Electronics*. Or could it be just resistance to change? Let's give them the benefit of the doubt!

continued on page 57

Logic design — 15

Action/status interface design

by B. Holdsworth* and D. Zissos† *Chelsea College, University of London

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The operation of action/status interfaces, which was the subject of part 14, continues, and design of interfaces is discussed, both in general and in two specific examples. This is the final article in the series on logic design.

With the exception of the go/no-go interface shown in Fig. 6, handshake system configurations which use two states to define the read/write cycle are easier to design and implement. Unless otherwise specified, the 'go' mode will start with a write operation. The diagram shown in Fig. 12(a) is used to define the read and write operations. For ease of reference the flip-flop used to generate the go/no-go signal G is shown in Fig. 12(b). The starting point in the design process is the basic system developed next.

Basic system

Read/write cycle. The implementation of a read/write cycle is straightforward. A block diagram of the two-device system is shown in Fig. 13(a), and its step-by-step operation is flow charted in Fig. 13(b). The state diagram of the interface logic is shown in Fig. 13(c).

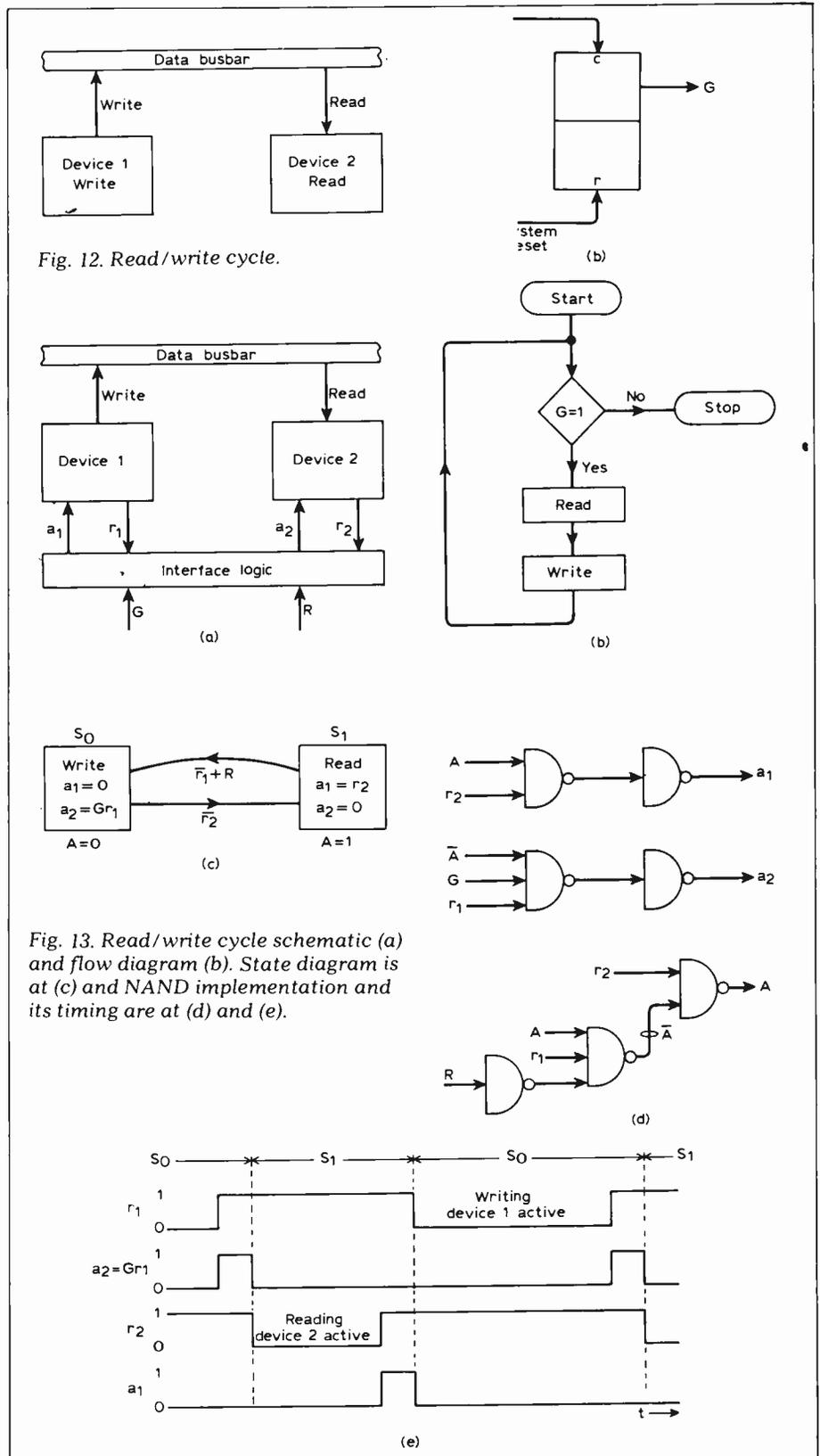
By direct reference to the state diagram the following equations are obtained:

$$\begin{aligned} \text{Turn-on set of } A &= \bar{r}_2, \\ \text{turn-off set of } A &= \bar{r}_1 + R, \\ A &= \bar{r}_2 = A r_1 \bar{R} \\ a_1 &= S_1 r_2 = A r_2 \\ a_2 &= S_0 G r_1 = \bar{A} G r_1 \end{aligned}$$

The NAND implementation of these questions is shown in Fig. 13(d).

The timing diagram for the read/write cycle is shown in Fig. 13(e). Initially, it is assumed that $r_1 = 0$ and that the system is in state S_0 , which implies that the system is writing and that device 1 is active. When device 1 has fully responded, r_1 goes from 0 to 1, $a_2 = G r_1$ and device 2 is activated. On activation its status signal r_2 goes from 1 to 0 and thus initiates the transition from S_0 to S_1 and also turns off the action signal a_2 .

In state S_1 the system is reading and it continues to do so until device 2 has completed its response, when r_2 changes from 0 to 1 and device 1 is activated. The cycle is completed when



the status signal of device 1, r_1 goes from 1 to 0, thus initiating a transition back from S_1 to S_0 and simultaneously turning off the action signal a_1 .

Write-inhibit cycle ($i_1 = 1$). To inhibit the write operation in the basic read/write cycle, the write operation is replaced by a read operation as shown in the flow chart of Fig. 14(a). The modification to the state diagram in Fig. 13(c) as a consequence of introducing the write-inhibit process is implemented by interchanging the values of a_1 and a_2 in state S_1 , when $i_1 = 1$. Expressed algebraically, the entries for write/inhibit are $a_1 = r_2\bar{i}_1$ and $a_2 = Gr_2i_1$, as shown in Fig. 14(b).

Read-inhibit ($i_2 = 1$). Similarly, to inhibit the read operation in the basic read/write cycle, a read operation is replaced by a write operation as illustrated in the flow chart of Fig. 15(a) and the corresponding modification to the state diagram is shown in Fig. 15(b).

The block diagram of the basic system with reset, go/no-go, read-inhibit and write-inhibit facilities is shown in Fig. 16(a). Its step-by-step operation is described by the flow chart of Fig. 16(b). The following equations are obtained directly from the state diagram shown in Fig. 16(c).

$$A = \bar{r}_2 + AR_1\bar{R}$$

$$a_1 = \bar{A}Gr_1\bar{i}_1 + Ar_2\bar{i}_1$$

$$a_2 = \bar{A}Gr_1\bar{i}_2 + AGr_2i_1\bar{i}_2$$

Design steps

The design of interfaces can be accomplished in the five steps listed below, and illustrated in Fig. 17. These steps are general and can be used in all interface design problems.

Aims of the design. The system specification is first expressed in the logic interface designer's terms. This step is introduced to ensure that the system requirements are interpreted correctly by the designer.

This stage is critical and requires cooperation between the interface designer and the system designer.

Device characteristics. In this step the designer specifies the terminal characteristics of the devices to be interfaced. Consideration of the purely internal characteristics of the devices should be avoided if possible.

System design. The interface designer specifies the system characteristics in general terms by means of a block diagram and a system flow chart and consults the designer for approval.

Hardware design. This step is provisional, and hardware design may well be modified as a consequence of the experience obtained in software design. It is accomplished conventionally using well-established methods described in this series.

Software design. On the basis of the hardware design and assuming the necessary instructions, the basic software for the operation of the device is designed. This process may well indi-

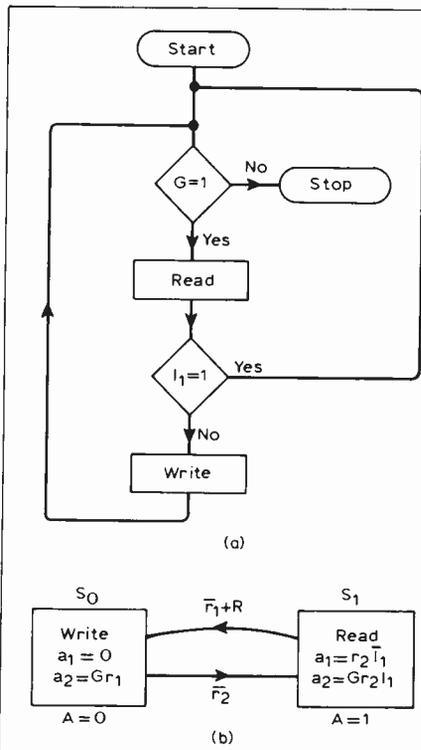


Fig. 14. Flow-chart for write-inhibit (a) and state diagram at (b).

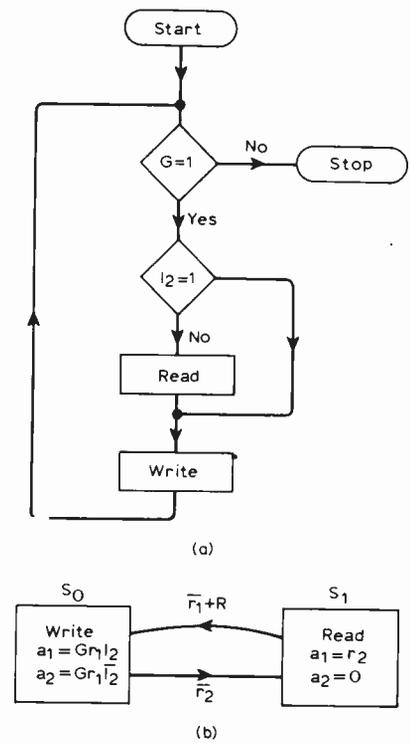


Fig. 15. Read-inhibit flow chart is at (a) and the state diagram at (b).

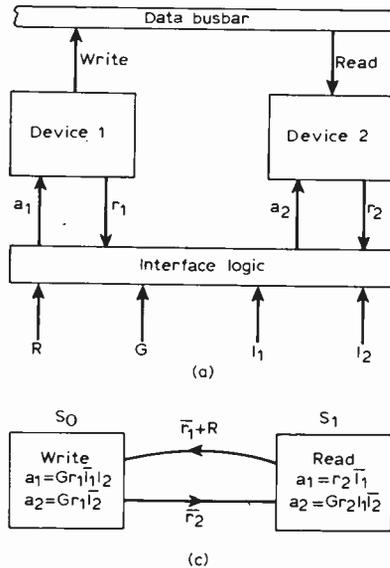


Fig. 16. (a) shows block diagram of basic system, with flow chart (b) and state diagram (c).

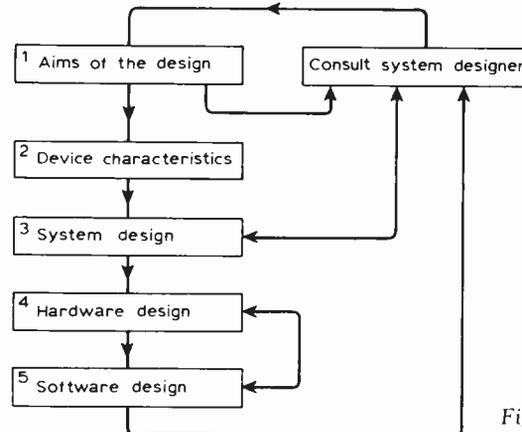
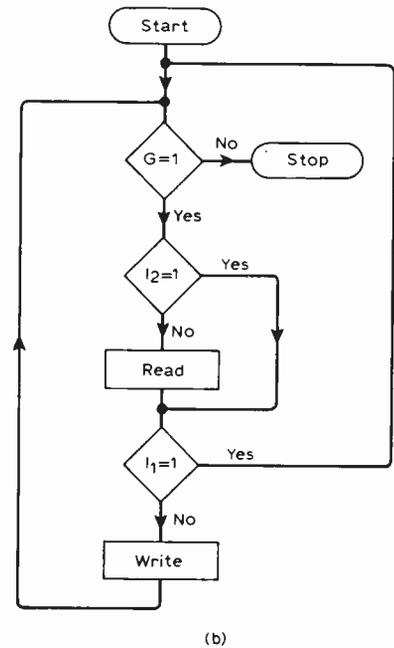


Fig. 17. Design steps for interface.

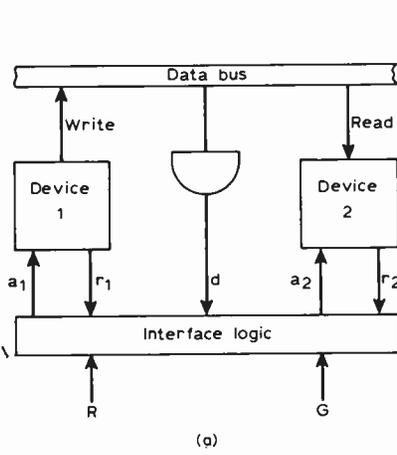


Fig. 18 (a) is block diagram for "rub-out characters" interface, with flow-chart at (b). State diagram is shown at (c) and circuit at (d).

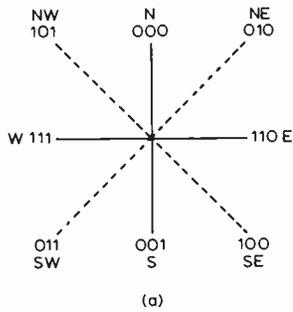
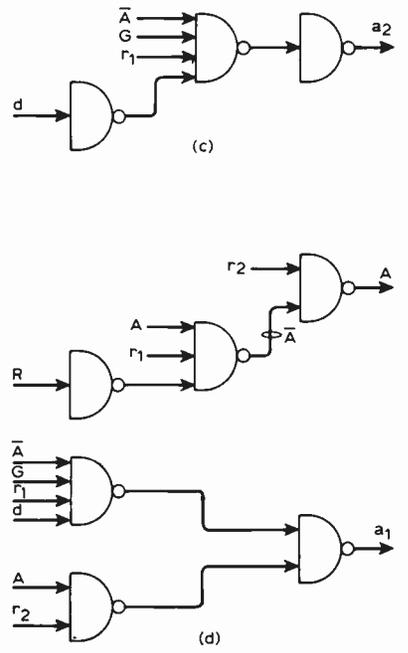
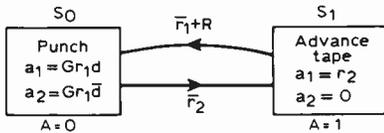
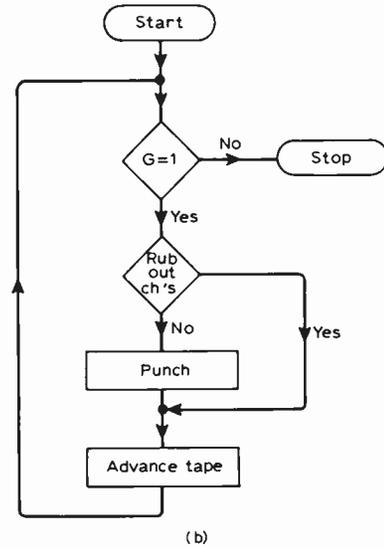
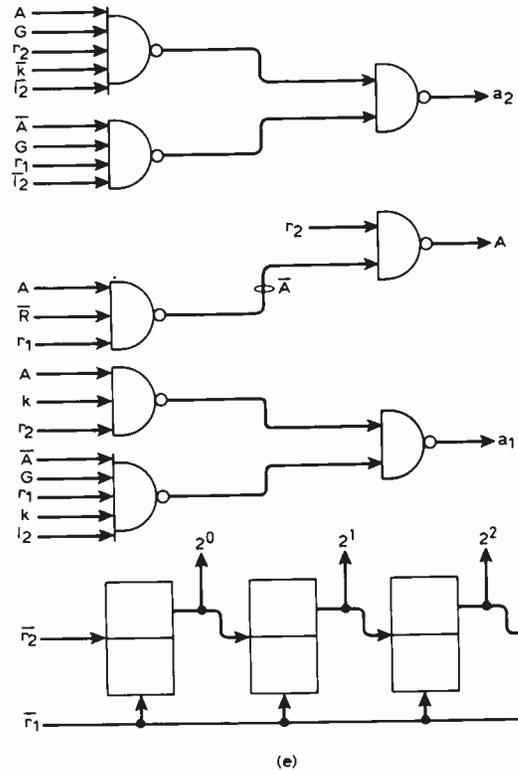
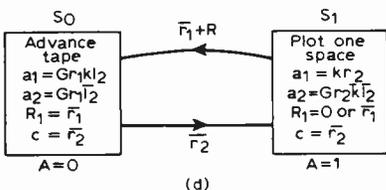
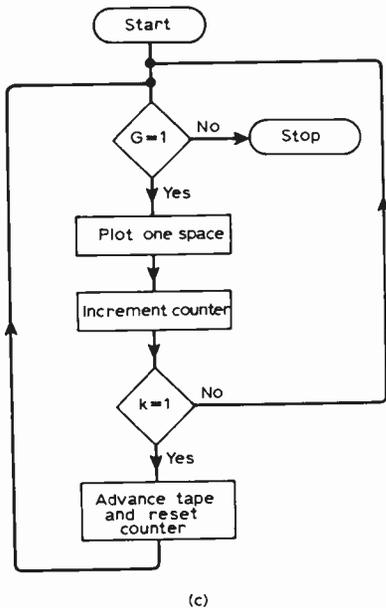
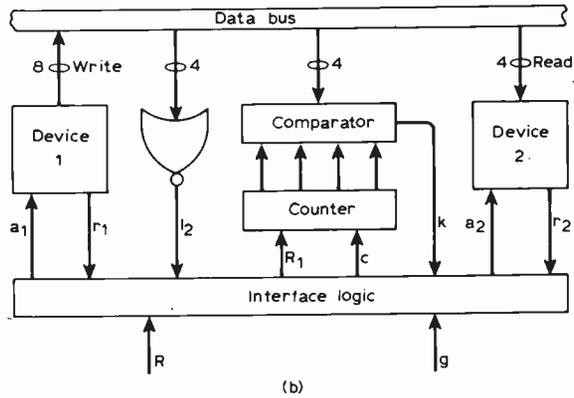


Fig. 19. (a) shows coding for eight directions of movement of digital platen, whose block diagram is at (b) and flow-chart at (c). State diagram and circuit are at (d) and (e).



cate modifications to the hardware design which may lead to improvements. In fact, software and hardware design should be regarded as complementary and should be repeated until a satisfactory design of both hardware and software has been achieved.

Problems and solutions

The design steps described are illustrated by means of two typical problems and their solution. For further design problems the interested reader is referred to the second edition of *Digital Interface Design*, by Zissos and Duncan, published by Oxford University Press, and to "System Design with Microprocessors," by Zissos, published by Academic Press.

Rub-out characters. Given a paper tape reader and a tape punch, design and implement a small system that allows a new tape to be produced in which the rub-out characters (all 1's) are deleted.

—The aim is to reproduce data after deleting specified characters, in this case the rub-out characters.

—Both reader and tape punch are action/status devices.

—The block diagram of the solution is shown in Fig. 18(a). The AND gate detects the rub-out characters on the data-bus. Its output d is logical '1' when all the digits on the data-bus are 1's. When $d=1$ the data is inhibited from being punched and the input tape is advanced. This is equivalent to $i_1=0$, and $i_2=d$ in the basic read/write cycle notation.

The flow chart describing the step-by-step operation of the system is shown in Fig. 18(b).

The state diagram of the interface can be derived either directly from the system flow chart in Fig. 18(b) or by substituting $i_1=0$ and $i_2=d$ in Fig. 16(c).

From the state diagram, which is shown in Fig. 18(c), the following equations are obtained:

$$\begin{aligned} A &= \bar{r}_2 + Ar_1\bar{R} \\ a_1 &= \bar{A}Gr_1d + Ar_2 \\ a_2 &= \bar{A}Gr_1d \end{aligned}$$

The implementation of these equations is shown in Fig. 18(d).

Reader-to-plotter interface. The first four tracks of an eight-track tape specify eight actions a digital plotter can take, namely move 0.1cm N, NE, E, SE, S, SW, W and NW with the stylus up or down. The other four tracks indicate the number of times each command is to be executed. For example 10010110 is interpreted as: "Draw a line 0.6cm long from NW to SE." Design a suitable interface between the reader and the plotter. The coding of the various directions specified in the problem is shown.

—The aim is as specified above.

—Both the reader and the plotter are action/status devices.

—The block diagram of the solution is shown in Fig. 19(b). In addition to the

two action signals a_1 and a_2 , the interface must reset the counter with signal R_1 and increment it with signal c at the appropriate times.

Initially the counter is cleared with the system reset signal R prior to the interface being activated by the signal G .

Activating the interface causes the pen to move one space (in this case 0.1cm.) in the direction specified, with the stylus up or down, as specified by the first four tracks of the tape. When the pen begins to move, r_2 becomes 0, and the counter is incremented. When the pen stops, indicated by r_2 becoming 1, the output of the comparator circuit shown in Fig. 19(b) is tested. If the output $k=0$, the pen is moved again and the counter is incremented. This continues until $k=1$, indicating that the stylus has moved through the number of distance units specified by the second set of four tracks on the tape. At this point the input tape is advanced and the counter is cleared. The process continues until the system is turned off, that is until $G=0$.

—As in the previous problem the state diagram can be derived directly from the flowchart of Fig. 19(c) and

$$i_2 = \sum_{b_4}^{b_7} b_i \quad \text{in the state diagram of}$$

the basic system. Note $i_2=1$ when the last four digits on the tape are zeros.

The modified state diagram is shown in Fig. 19(d) and by direct reference to this figure the following equations are obtained:

$$\begin{aligned} A &= \bar{r}_2 + A\bar{R}r_1 \\ a_1 &= \bar{A}Gr_1kL_2 + Ak_r_2 \\ a_2 &= \bar{A}Gr_1\bar{L}_2 + AGr_2\bar{k}i_2 \\ R_1 &= \bar{r}_1 \\ c &= \bar{r}_2 \end{aligned}$$

The implementations of these equations is shown in Fig. 19(e).

Acknowledgements

The authors are grateful to Mr J. Bothoroy, a research assistant at the University of Calgary, for his contribution towards the development of action/status interfaces.

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- "Digital Interface Design," D. Zissos and F. G. Duncan, Oxford University Press 1974.
- Microprocessor System Design," D. Zissos, Academic Press 1978.

□

SIXTY YEARS AGO

In late 1918, the triode (a name not yet used) was beginning to gain ground, and several valve receivers were in use. But its acceptance was not as rapid as it might have been, particularly under the stimulus of war-time development. A note appeared in this issue on the Marconi Double Note Magnifier which evidently exhibited a gain of three times per stage.

"Briefly, the instruments consist of two three-electrode valves connected in series with one another in such a way that the telephone currents from the magnetic detector or crystal receiver are magnified in two successive stages before being led to the telephones themselves. The valves thus act not as detectors but as amplifiers. All circuits have been simplified to the highest degree so as to remove the need for adjustments. Electrically there is no difference between the model for the magnetic and that for the crystal except in the design of the first transformer, the primary for which has, of course, to be of lower resistance for the magnetic detector than for the crystal. In one model a switch takes the place of three sets of terminals, and is connected in such a way that, when working direct without magnification, the valve filament circuits are broken. When using first magnification, one valve only is in circuit, whilst for second magnification both valves are in circuit. Otherwise the arrangements are the same.

The total arrangement obtained with this new instrument is such that signals from the magnetic detector are at least three times as strong as those obtainable with a crystal receiver. It will be noticed that a 200-volt battery is used for the plate circuit.

By the addition of the note-magnifier to the magnetic, we have available a receiver which possesses the notable reliability of the magnetic, and far greater sensitiveness than the crystal, which — with the exception of the more complicated forms of valve receivers — has hitherto formed the most sensitive commercial type."

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The f.e.t. as detector

Improvements in performance over the diode detector

by Roger S. Amos, B.Sc.

The author sets out his views on the advantages of field-effect transistors over semiconductor diodes as demodulators for a.m. receivers. Reduced distortion and improved signal-to-noise ratio are claimed and a design for a receiver using the technique is presented.

MOST MODERN radio and television receivers use silicon or germanium diodes to demodulate the output of the i.f. amplifier. Such detectors are purely passive, giving no power gain, but in superhet receivers this is generally of no account, since the r.f. and i.f. stages provide all the sensitivity that is needed and deliver the power to drive the detector. Before the advent of the superhet, however, the detector in a t.r.f. receiver was often a triode valve which gave a.f. amplification besides detection, contributing to the sensitivity of the receiver.

Today, with the ready availability of f.e.t.s, which in many respects behave like triode valves, it is possible to apply the advantages of semiconductor technology to the triode detector circuits of yesteryear. And this is no purely academic matter, for some popular a.m. superhet i.cs, such as the NE546A, LM1820N, μ A720 and CA3123E (which are all pin-for-pin equivalents) and the rather different TBA651, have an uncommitted i.f. output, leaving the manufacturer or constructor to add the detector of his choice.

The application of f.e.t.s as a.m. detectors, however, demands an understanding of both f.e.t. parameters and triode detector operation. The latter may not be familiar to those who have learned radio theory in the age of the superhet and especially the semiconductor superhet. The three most popular triode detectors were the leaky-grid detector, the infinite-impedance detector and anode-bend detector. But when these are simulated using f.e.t.s, unexpected results are often obtained. This article seeks to explain those results and to show how the f.e.t. may provide an attractive alternative to the diode as an a.m. detector.

Leaky-grid (or-gate) detector

The leaky-grid detector, a typical circuit of which is shown in Fig. 1, is essentially an extension of one form of thermionic

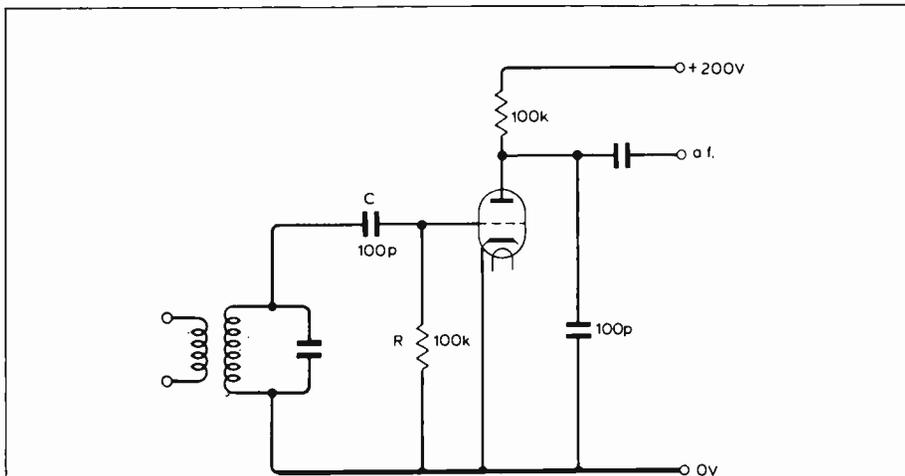


Fig. 1 A basic leaky-grid detector. The component values shown are typical.

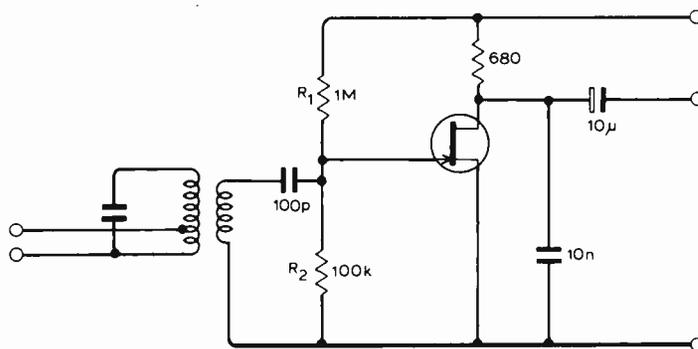


Fig. 2 Practical circuit for the f.e.t. equivalent of the leaky-grid detector shown in Fig. 1. Component values are typical.

diode detector, the grid of the triode acting as the anode of a diode. Hence the RC network between tuned circuit and valve are similar to those used in diode circuits, and for high-quality a.m. detection would be chosen to give a time-constant of say 50μ s. Positive half cycles of input cause grid current to flow (hence the circuit's name) rapidly charging the reservoir capacitor C. This charge leaks away slowly through resistor R, because this is high compared with the forward resistance of the "diode" when charging. Thus, before the charge has leaked away it is restored by the next positive half cycle. The voltage across the capacitor there-

fore follows the modulation of the incoming carrier and because this voltage is applied to the valve between grid and (through the tuning inductor) cathode, the circuit gives both detection and audio amplification.

An f.e.t. equivalent of this circuit would probably employ a junction-gate f.e.t. since, in these, sufficient positive bias on the gate (an n-channel device is assumed) drives current across the gate-channel junction. Insulated-gate f.e.t.s in normal use would be unsuitable since gate current is precluded, but a device having a separate substrate or base terminal could possibly be used with that terminal as the "grid"; if the

device is an enhancement-type, the gate would need to be biased forward to establish drain current.

Direct substitution of the valve in Fig. 1 by a suitable j.u.g.f.e.t. with appropriate amendment of component and supply values generally yields disappointing results, the output being feeble and distorted. The principal reason for this is that the gate-channel junction, like a silicon diode, does not conduct appreciably if the forward voltage is below about 0.6V. When silicon diodes are used as a.m. detectors, it is regular practice to apply forward bias to them to bring them into conduction, while maintaining the working point on a sufficiently non-linear part of the characteristic to achieve the required detection; often this bias is provided through the a.g.c. loop in the receiver. The addition of positive bias to the gate of the f.e.t., as shown in Fig. 2, greatly improves the audio quality and efficiency of detection. In the valve circuit of Fig. 1 no positive grid bias was needed because some electrons are emitted from the cathode with sufficient energy to land on the grid even when it is slightly negative relative to the cathode.

There are two main advantages in this form of detector. Firstly, like the diode detector on which it is based, it imposes a load on the tuned circuit through which power is delivered to drive the grid-cathode or gate-channel "diode". This loading can cause distortion since the "diode" resistance is non-linear, falling with increasing positive half-cycle amplitude. In addition there is some steady damping of the tuned circuit through R in Fig. 1 and the effectively parallel R_1 and R_2 in Fig. 2. These reduce the Q of the tuned circuit, limiting the gain available from the previous stage. In superhet receivers, of course, the slight power loss is no great problem since sensitivity and power are available from the preceding i.f. amplifier; in early t.r.f. receivers the loss of power, sensitivity and also selectivity was less tolerable. In high-quality applications, however, the distortion may be less acceptable. Damping of the tuned circuit can be counteracted in some measure by driving the detector from a tapping or a secondary winding, although this further reduces the power available.

Secondly, because the "diode" must be capable of giving appreciable conduction on positive half cycles of input, no reverse bias may be applied to it, hence the absence of cathode and source resistors in the circuits of Figs. 1 and 2. Consequently current consumption is heavy, the f.e.t. giving best results as its saturation drain current is approached; this can be as high as 20mA for a 2N3819, 15mA for a BF244B and 13mA for a BF256LB. While this is of little consequence in mains-driven equipment, it is clearly wasteful in battery equipment. The detector

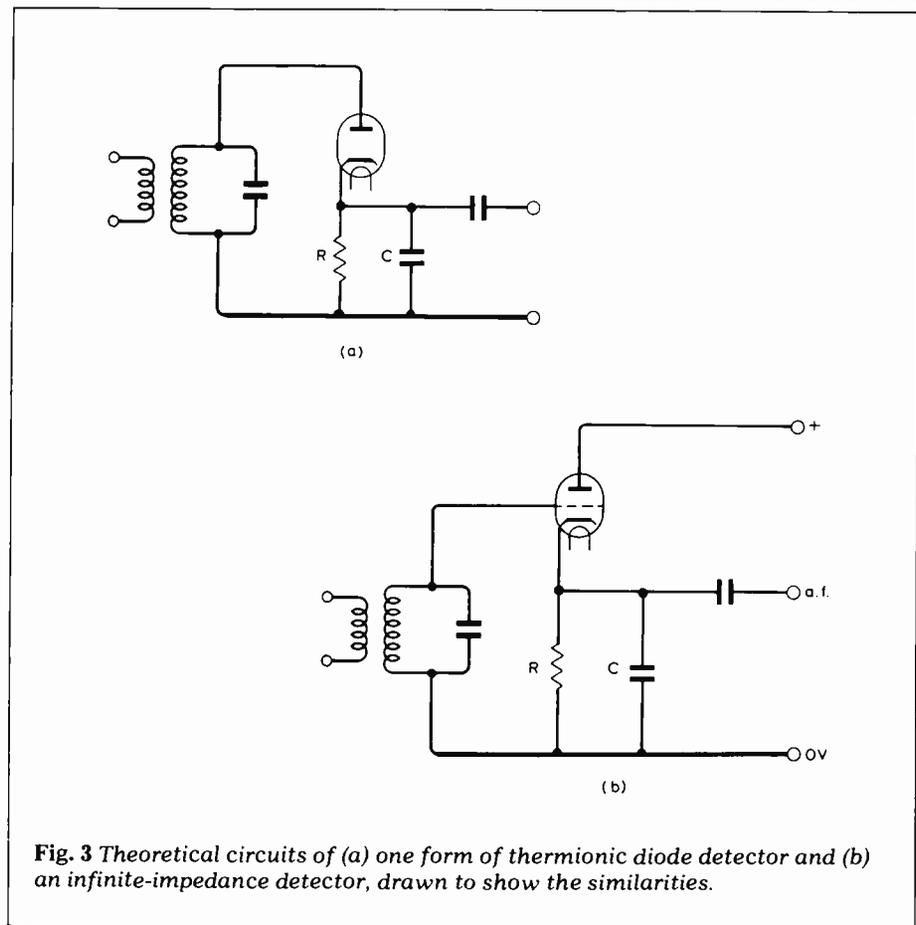


Fig. 3 Theoretical circuits of (a) one form of thermionic diode detector and (b) an infinite-impedance detector, drawn to show the similarities.

circuits to be described avoid both these disadvantages.

Infinite-impedance detector

As Fig. 3 shows, the infinite-impedance detector closely resembles one form of thermionic diode detector. The difference is that in the former the power to drive the detector comes from the supply rather than the input tuned circuit. This gives it two advantages over the leaky-grid detector. Firstly, anode current in the cathode load resistor provides grid bias, ensuring that an effectively infinite impedance is presented to the input tuned circuit. This minimizes damping and enables the detector to make full use of the available input voltage. Consequently, distortion is low and at one time this detector was favoured in high-quality applications. Secondly, the cathode resistor limits anode current, which may be very low. Indeed, if the valve is near cut-off the non-linearity of its grid bias/anode current characteristic enhances the efficiency of detection; this non-linearity is not, however, essential to the operation of the infinite-impedance detector.

Positive half cycles of input voltage cause peaks of anode current which rapidly charge the reservoir capacitor. Since the only discharge path presents a comparatively high resistance, the charge leaks away slowly. Consequently, negative half cycles of input voltage oppose the charge on the reservoir capacitor, causing the valve to

be biased back. Providing the cathode resistor and capacitor are chosen to give a suitable time constant, the cathode voltage will accurately track the peaks of the positive excursions of input voltage. Since the detector is a cathode follower, it resembles a diode circuit in that it gives no voltage gain; it does, however, give appreciable power gain in that it transfers voltage from its practically infinite input impedance to the comparatively low output impedance at the cathode.

The valve in Fig. 3(b) may be directly substituted by an f.e.t. Since the gate will not be required to conduct, insulated-gate and junction-gate types are equally suitable, but enhancement-mode devices would require more complex biasing arrangements. Since junction-gate types are inexpensive and easier to handle, they will probably be preferred.

Figure 4 shows an infinite-impedance detector using an n-channel j.u.g.f.e.t. The circuit generally gives an excellent signal-to-noise ratio, low distortion and a higher level of audio output than might be expected; the reason for this will be discussed below. Mean drain current rises as signals are received and if amplified can be used to derive a.g.c. or "S" meter drive. In many respects this is the most promising f.e.t. detector.

The circuit, has, however, a disadvantage. In an f.e.t. there is some capacitance between gate and source; this is given as typically 8pF for the 2N3819 and 4pF for the BF244 and

BF256. Fig. 5 shows the circuit of Fig. 4 including this capacitance and a decoupling capacitor redrawn to show its similarity to a shunt-fed Colpitts oscillator. If the f.e.t. is a high-gain type or if it has a high gate-source capacitance, instability may occur. This can generally be eliminated by making the source resistor large (and the reservoir capacitor correspondingly small) thereby reducing the gain of the f.e.t. In practice it is often helpful to make the source resistor a pre-set pot. (50kΩ is a useful value) which can be adjusted for stability and the reservoir capacitor (470pF is a typical value) can be replaced by a more suitable value if necessary. If the detector is fed from a damped tuned circuit or from a voltage step-down transformer (as in Fig. 8) instability is unlikely to cause any problems.

There are circumstances in which the potential instability of this detector can be an advantage. In receiving c.w. and s.s.b. transmissions it is often helpful if the detector can be made regenerative, and this can often be achieved very simply by the addition of a suitable external capacitor between gate and source. A variable source resistor will usually give smooth control of regeneration. If the f.e.t. is a low-noise type such as the BF256, the detector alone at the threshold of regeneration exhibits remarkable sensitivity; one constructed by the author having only a ferrite aerial and followed by a low-noise audio stage feeding a power amplifier equalled several domestic superhet receivers in sensitivity, but gave better signal-to-noise ratio and less distortion. Almost certainly the inherent positive feedback through the f.e.t.'s internal gate-source capacitance and associated external circuitry accounts for the unexpectedly high level of audio output.

Anode- (or drain-) bend detector

Like the infinite-impedance detector, the anode-bend detector, of which a typical circuit is shown in Fig. 6, presents an infinite impedance to the input tuned circuit. Detection is, however, by virtue of the non-linear characteristics of the valve near cut off. Thus, this detector shares with that previously described the advantages of minimal damping and low current consumption; furthermore, it offers the extra advantage of useful audio amplification.

The operation and design of this detector are most easily understood if it is regarded as a stage of audio amplification in which the input is at r.f. If the valve were perfectly linear and if the input signal were too small to cause overloading there would, of course, be no audio output. But because the valve is biased nearly to cut-off and is non-linear the increase in anode current on each positive half cycle of r.f. input is greater than the diminution caused by an equal negative half cycle and the

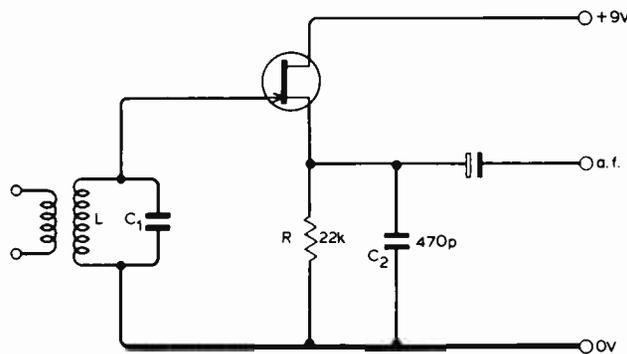


Fig. 4 An infinite-impedance detector using an n-channel j.u.g.f.e.t. The component values are typical, but there are complications – see Fig. 5 and text.

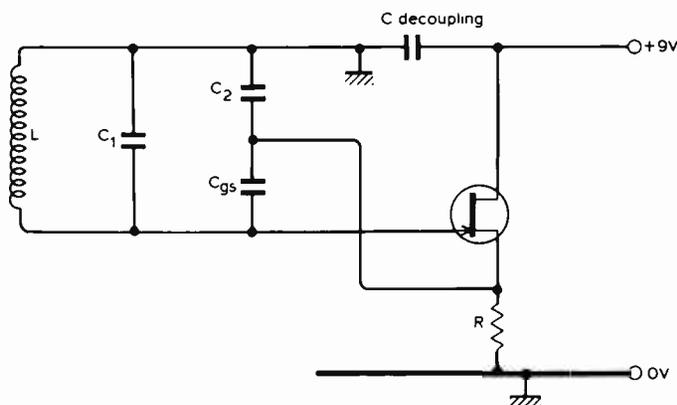


Fig. 5 Part of the circuit of Fig. 4 including the internal gate-source capacitance of the f.e.t. (C_{gs}) and a decoupling capacitor across the supply ($C_{decoupling}$) redrawn to show the similarity of the infinite-impedance detector to a shunt-fed Colpitts oscillator. This explains the instability sometimes encountered in this form of detector.

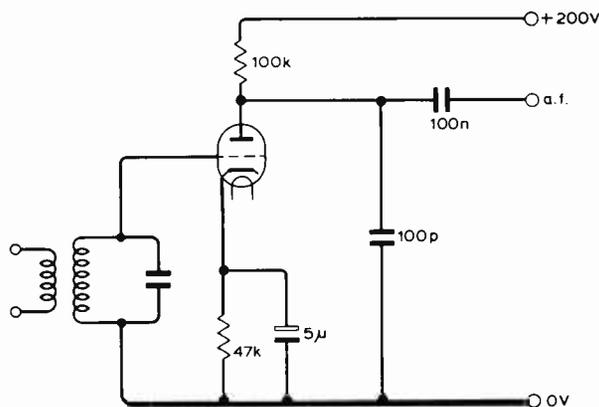


Fig. 6 Theoretical circuit of an anode-bend detector. The component values are typical.

mean anode current follows the modulation of the incoming carrier. This is converted to a voltage output by the insertion of an RC combination with a suitable time constant in the anode circuit of the valve. The author's experiments (with f.e.t. circuits) have shown that the cathode (or source) resistor needs to be decoupled at audio

frequencies, because it is common to input circuits and, without decoupling, introduces negative feedback, reducing the audio gain. With a source resistor of 47kΩ the optimum bypass capacitor is about 10μF; higher values cause distortion and lower ones upset the audio frequency response.

The transfer characteristics of f.e.t.s

resemble the grid voltage/anode current characteristics of triode valves, their non-linearity near cut-off suiting them theoretically for use in "drain-bend" detectors. As in the infinite-impedance detector, the gate will not be required to conduct, so that junction-gate and insulated-gate types are equally suitable. If an enhancement-mode device were used, it would be necessary to hold the gate bias just above the threshold voltage, which could cause complications.

If an f.e.t. were substituted for the valve in Fig. 6 and the component and supply values amended accordingly, the performance of the resulting drain-bend detector would probably be disappointing compared with that of the simpler infinite-impedance detector shown in Fig. 4. In fact the drain-bend detector poses a number of problems.

Firstly, there is a loss in sensitivity caused by the Miller effect, which consists of negative feedback at radio frequencies through both internal and stray drain-gate capacitance. Although great losses might be expected because of the high value of drain load resistance and small drain bypass capacitor,

this is largely counteracted by the low g_m (or y_{fs}) of the f.e.t. near cut-off. Miller effect can be overcome by the use of neutralization or a cascode circuit; a dual-gate i.g.f.e.t. can be employed since it behaves like a cascode. Although the helpful regenerative tendency seen in the infinite-impedance detector is destroyed by the high drain load resistance and extensive source decoupling, it can be re-introduced by partial decoupling at the source, as shown in Fig. 7. At the sacrifice of some audio voltage gain this permits the introduction of an external capacitor between source and gate to neutralise Miller effect, improving the sensitivity. It also facilitates regeneration if required. The resulting detector, however, is no longer a pure drain-bend detector — it is a hybrid between drain-bend and infinite-impedance types.

Secondly, the output impedance at the drain may be as high as 50k Ω . If this is coupled to a common-emitter audio stage having a low input impedance (2k Ω is typical), the detector output voltage will collapse, leading to apparently poor results. For this reason it is advisable to follow a drain-bend

detector with either an emitter-follower, as in Fig. 7, or an f.e.t. audio stage.

Thirdly, a high supply voltage may be needed. With the f.e.t. operating near cut-off, the voltage across the source resistor approaches the pinch-off voltage. If the f.e.t. were a 2N3819 this may be as high as 5V, and if the drain load resistor were, say, four times the source resistor, clearly 20V would be needed to drive the drain current through it. Ideally, at least 3V should be maintained between drain and source so that the minimum supply voltage under these conditions is 28V. This can be overcome by deliberately selecting an f.e.t. with a low V_p ; some BF244s have V_p less than 1V and, using these, drain-bend detectors can be constructed which will operate satisfactorily from 9V and even 6V supplies. The BF256 can also be used with a 9V supply if component values are chosen carefully.

Fig. 7 shows a practical circuit for a hybrid drain-bend/infinite-impedance detector with emitter follower output stage in which all the performance-saving steps outlined above have been

Fig. 7 Practical circuit diagram for a hybrid drain-bend/infinite-impedance detector with buffer emitter-follower audio stage; the latter is necessary to match the high output impedance of the detector. Neutralization is applied via the 25pF trimmer to counteract Miller effect. But the performance may in some respects be inferior to that of the simpler infinite-impedance detector in Fig. 4.

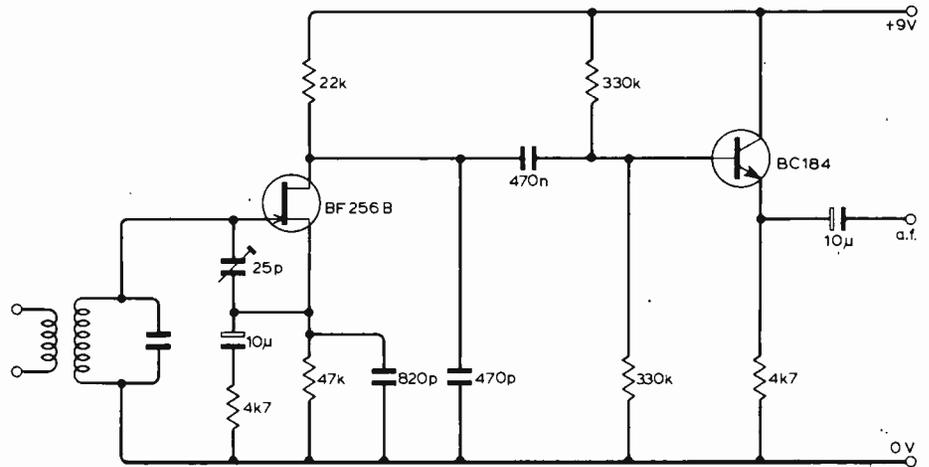
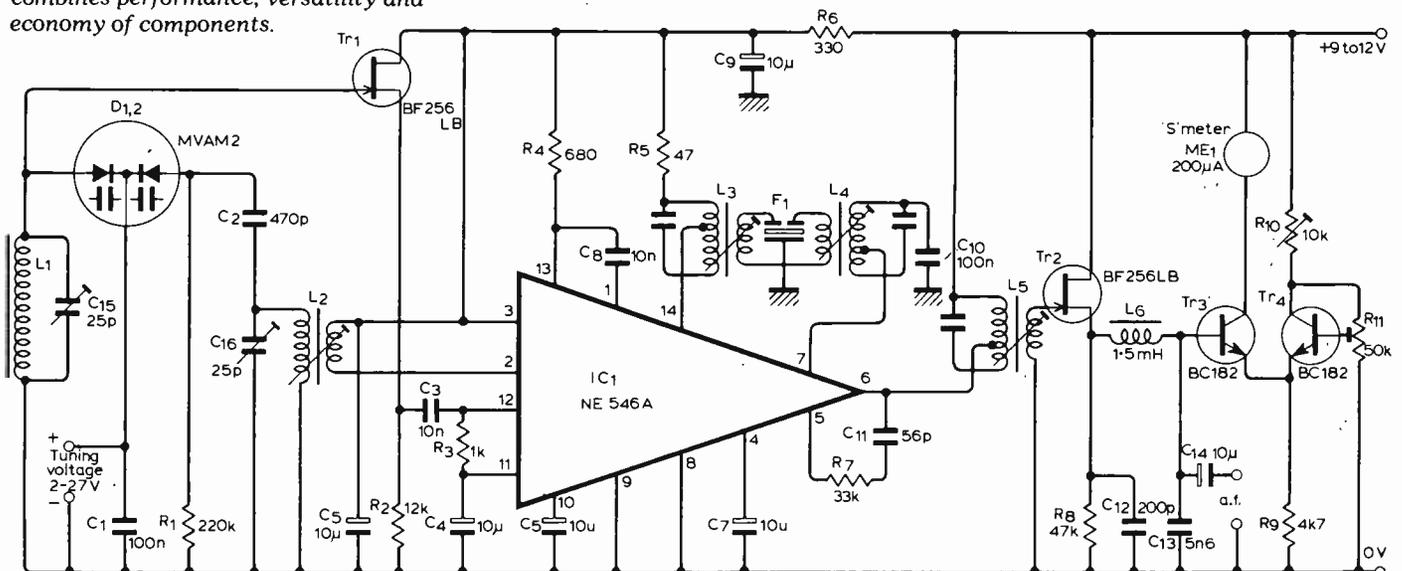


Fig. 8 Circuit of an a.m. superhet using an i.c. for main functions and an f.e.t. infinite-impedance detector. The circuit combines performance, versatility and economy of components.



The author

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taken. While it will generally give a greater audio output than its counterpart in Fig. 4 from the same input signal, the signal-to-noise ratio may well be inferior, and the circuit may suffer from treble-cut caused by the extra positive feedback introduced to neutralize the deprecations of Miller effect. The infinite-impedance detector shown in Fig. 4 followed by a stage of low-noise a.f. amplification would give results superior in most respects using almost the same components. For this reason the latter circuit will probably be preferred.

Complete receiver

Figure 8 shows the circuit of a superhet receiver in which a readily available integrated circuit provides the functions of r.f. amplifier, local oscillator, mixer, i.f. amplifier and a.g.c. detector, an infinite-impedance detector like that in Fig. 4 providing a.f. output and a d.c. feed for an "S" meter circuit. Although a two-gang mechanical tuning capacitor could be used, the author's prototype employed a Motorola MVAM2 dual varicap diode, the tuning voltage being selected from an array of seven pre-set and one variable 100k Ω pots by an eight-way push-button switch; the 27V bias was provided by three 9V dry batteries in series, which proved remarkably stable, the pre-set stations remaining in tune over many weeks.

Source-follower Tr_1 matches the aerial tuned circuit to the r.f. stage and, helped by inherent regeneration as in an infinite-impedance detector, contributes to the sensitivity and signal-to-noise ratio of the receiver. The combination R_4/C_8 provides broadband coupling between r.f. stage and mixer. Coupling from mixer to i.f. amplifier is via the selectivity block $L_3/F_1/L_4$ which may consist either of two discrete i.f. transformers and a ceramic filter or, as in the author's prototype, an integrated block containing these elements; these are available commercially with a choice of bandwidths. The i.f. output appears at pin 6, a portion being fed through C_{11}/R_7 to a voltage doubling diode pair in the i.c. which provides a.g.c. for the r.f. stage. Signal is transferred from i.f. amplifier to detector through a discrete i.f. transformer, which in the author's prototype gave a

voltage step-down, being intended for coupling to a diode detector. Although this transformer was far from ideal for its present detector, the detector nevertheless gave more than adequate output to feed a domestic high-quality amplifier. The step-down transformer also improved the stability of the detector. The combination L_6/C_{13} removes any stray i.f. which might cause distortion in the amplifier or "S" meter circuit.

The optional "S" meter drive consists of the long-tailed pair $Tr_{3/4}$ and associated circuitry. Mean drain current in the infinite-impedance detector rises when a signal is received; this appears as an increased voltage across the source resistor R_8 , which biases Tr_3 forward; in the absence of a signal Tr_3 is biased off by Tr_4 . For meters up to 500 μ A f.s.d. R_{10} and R_{11} can be adjusted for meter sensitivity and zero respectively; meters over 500 μ A f.s.d. may be satisfactory if R_9 is reduced. Additional a.g.c. could be derived from Tr_3 collector by the inclusion of a suitable resistor in series with the meter; this could be fed to the gate of Tr_1 , the signal being fed in through a suitable capacitor.

Sensitivity of the receiver compares favourably with that of domestic superhet receivers; signal-to-noise ratio is superior and distortion very low. The latter two features are due in some measure to the nature of the detector. With junction-gate f.e.t.s costing little more than silicon diodes, the infinite-impedance detector surely offers scope for an improvement in a.m. receiver design. \square

LITERATURE RECEIVED

"Glossary of Microelectronic Terms, Definitions and Symbols" — JEDEC publication No. 99 costs \$7.50 and is obtainable from Standards Sales Office, Electronic Industries Association, 2001 Eye Street, N.W., Washington, D.C. 20006, U.S.A.

Publications of the International Telecommunications Union are listed in a catalogue available from General Secretariat, I.T.U., Sales Service, Place des Nations, CH-1211 Geneva 20, Switzerland WW401

Electrostatic printer/plotter S1004, in both alphanumeric and graphic forms, described in a brochure from EMI Technology (SE Labs), Data Products Division, Spur Road, Feltham, Middlesex TW14 0TD. ... WW402

Personal computer for small business use described by ITT in a leaflet from ITT Consumer Products (UK) Ltd, Chester Hall Lane, Basildon, Essex WW403

Transducer energising and conditioning is the function of Industrial Unit IG2104, described in a leaflet obtainable from Carl Schenck (UK) Ltd, Stonefield Way, Ruislip, Middlesex HA4 0JT WW404

Transient suppressor diodes, from 6.8 to 200V, are the subject of a data sheet from Semtech, which can be obtained from Bourns (Trimpot) Ltd, Hodford House, 17/27 High Street, Hounslow, Middlesex WW405

Miniature motors for 1978 from Portescap, using ironless motors for low inertia, described in a leaflet from Portescap (UK) Ltd, 204 Elgar Road, Reading RG2 0DD WW406

Antennas and masts are fully detailed in a large catalogue from American Electronic Laboratories, Inc., Dept. 1122, P.O. Box 552, Lonsdale, Pennsylvania 19446, U.S.A. WW407

Timers working in the 'delay on de-energize' mode, are made by Tempatron and are covered by leaflet TDD9/77 from Tempatron Ltd, 6 Portman Road, Reading WW408

Microphones, headphones, communications and test instruments are listed and fully detailed in the Sennheiser 'Revue 9' catalogue. Contains much useful information — in English. Hayden Laboratories Ltd, Hayden House, Churchfield Road, Chalfont St. Peter, Bucks SL9 9EW WW409

Screwdriving equipment, including power bits, screwdriver kits and adaptors are illustrated in Catalogue 3a from Harmsworth Townley & Co. Ltd, Todmorden, Lancs OL14 5JY WW410

Audio accessories catalogue, which also presents information on multimeters and digital timepieces, obtainable from Ross Electronics, 32 Rathbone Place, London W1P 1AD WW411

Company analysis, considering the performance of 50 largest firms in European electronics over the last few years, is obtainable at £18 from Mackintosh Publications Ltd, Mackintosh House, Napier Road, Luton LU1 1RG.

Frequency synthesizers — 1

The generation of wanted frequencies from other frequencies

by R. Thompson, M.I.E.E.

The term "frequency synthesis" is applied to processes involving the generation of some wanted frequency from one or more other frequencies. The most common forms of frequency synthesizer use high-grade fixed frequency references to generate fixed- or variable-output frequencies with stabilities similar to those of the references. Like many definitions, this one cannot be considered to be precise and we shall see that frequency synthesis represents a particular grouping of techniques, most of which are widely used in other applications of electronics.

THE NEED to generate variable frequencies with the stability of a fixed reference has led to a concentration of effort on frequency synthesis over the past 15 years. A major application of this type of synthesizer is in radio communication equipment where narrow-band modulation methods with precise carriers are required. Another major application has been in modestly priced instruments capable of very accurate

measurement of time, frequency and phase.

A feature of great importance with many modern synthesizers is the ease with which the required frequency can be selected. Communication equipment design has had increasing emphasis on ease of operation, aiming in many cases to eliminate the need for specialist operators. Nowhere has this pressure been greater than with military applications. Here, the trend has been from equipment requiring tediously repetitive adjustment to switch selectable operation, and now to radio equipment having entirely automatic frequency control.

The basic requirement for generating one frequency from another can be stated as: $f_2/f_1 = X/Y$, where X and Y are rational numbers. In principle, therefore, synthesis only requires multiplication (X) and division (Y). However, as we shall see, practical considerations limit the attainable values of X and Y , where such practical limitations occur

X/Y can be factored as: $X/Y = (x/y)(X_1 \pm X_2/Y_2)$. The introduction of the \pm allows the multiplication/division factors to be reduced. We shall be coming back to this in more detail later; the important point here is to see that we are interested in the four basic arithmetic functions.

Addition and multiplication

If we start with some very simple waveforms and their spectra we can get an appreciation of the possibilities and difficulties of adding and multiplying frequencies.

Figure 1 shows simple rectangular waveforms and their associated frequency spectrums. Frequencies are only present at integral multiples of the fundamental frequency f_1 and, from the general expression for the Fourier series shown below, it can be shown that their amplitudes follow a $\sin x/x$ law.

$$a(t) = \frac{A\tau}{T} \left[\frac{1}{2} + \sum_{m=1}^{\infty} \frac{T}{\tau m\pi} \sin\left(\frac{\tau m\pi}{T}\right) \cdot \cos\left(\frac{2\pi m t}{T}\right) \right]$$

Obviously a wide range of frequencies can be generated, but if a reasonably flat spectrum is required up to high order harmonics, very narrow pulses will be required. This will, however, result in very little energy being available at any selected frequency.

If we are interested only in a particular range of harmonics, one method is simply to filter the output of the square wave, even though the level quickly reduces as we increase the harmonic number. An alternative method is to use the squarewave to switch a sinusoidal signal frequency f_2 , as shown in Fig. 2.

In Fig. 2(a) there is an integral relationship between f_1 and f_2 , and each burst of f_2 starts in the same phase. The resultant spectrum is a double-sided version of that in Fig. 1 centred on f_2 .

In Fig. 2(b), where $f_2 \neq m f_1$, the squarewave not only switches f_2 on and off but forces it to start each burst in the same phase. This gives a spectrum similar to that in Fig. 2(a) but now, although the spectral envelope is centred on f_2 , there is no component there. Since the waveform is periodic at f_1 the components will be at harmonics of f_1 . The forced synchronisation of an oscillator, producing an output as in Fig. 2, gives high level outputs with a frequency stability dependant only on f_1 and it can be seen that by controlling

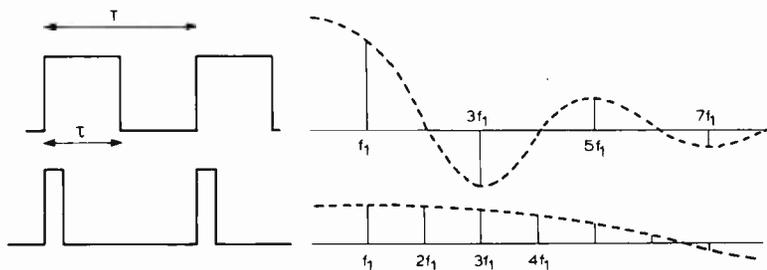


Fig. 1. Simple rectangular waveforms and graphical representations of their associated frequency spectrums.

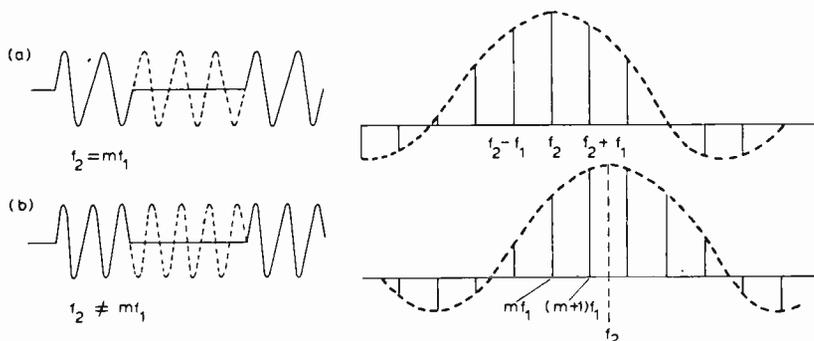
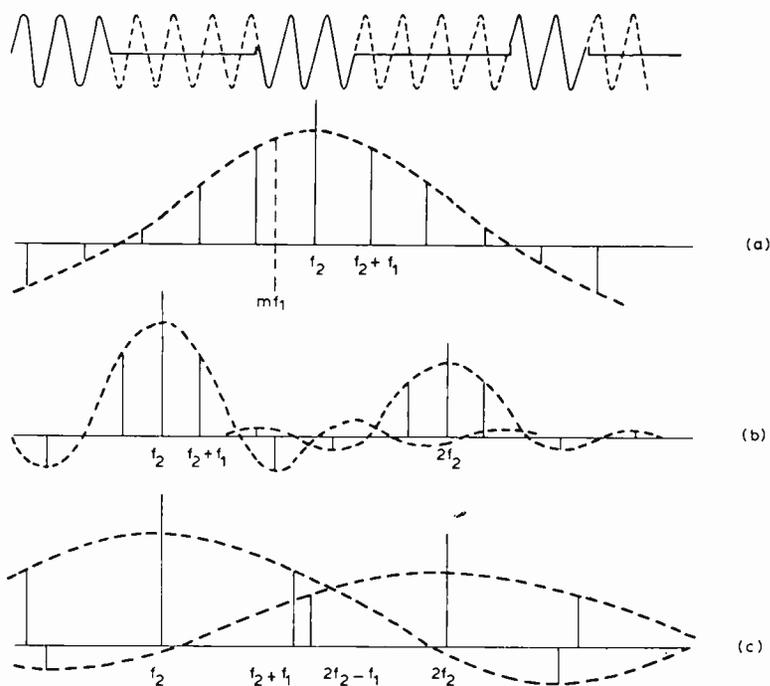


Fig. 2. A squarewave f_1 being used to switch a sinusoidal signal f_2 to produce frequency spectrums. In (a) each burst of f_2 starts in the same phase. In (b) the squarewave forces each burst of f_2 to start in the same phase.



◀ **Fig. 3.** Diagram showing three frequency spectrums resulting from a mixing process which enables frequencies to be added or subtracted. (a) is probably a simplification because it would normally also contain harmonics of f_2 . In (b) the ratio of f_2/f_1 is high and $f_2 + f_1$, f_2 and $f_2 + 2f_1$ are difficult to separate. In (c) the problem is that of removing $2f_2 - f_1$.

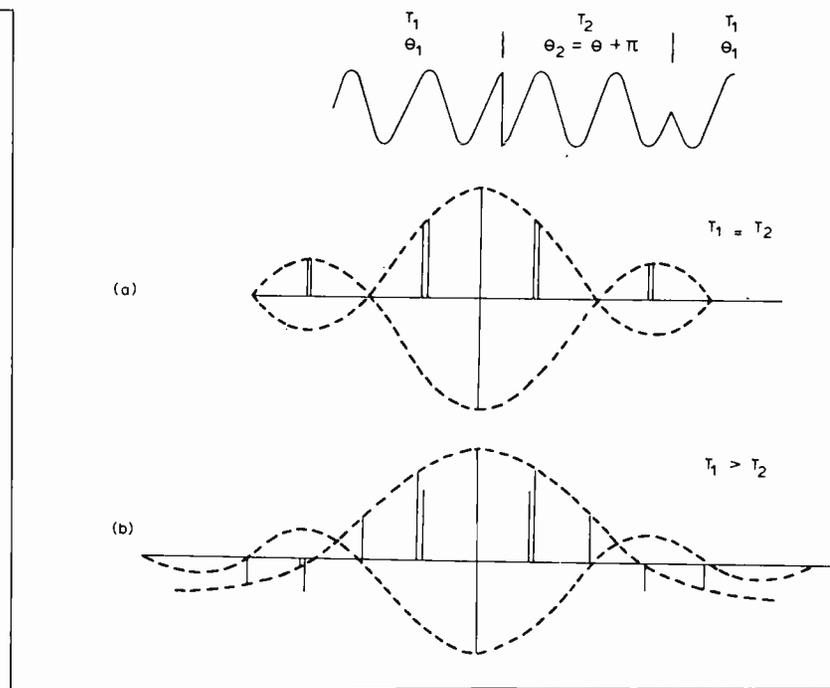
Fig. 4. The use of equal on and off periods in Fig. 3 enables the $f_2 + 2f_1$ frequency component to be suppressed. This diagram shows how, by also replacing the off period with a phase-reversed f_2 , f_2 may be suppressed too. In (a), where the periods are equal, the nearest components are $2f_1$ away. (b) shows what can happen when the periods are not equal.

the switching pulse width a form of filtering is provided.

One of the recurrent problems in frequency synthesis is the selection of the wanted signal from a host of unwanted signals and any method capable of providing this selection is of interest. The most obvious methods are L,C,R and crystal filtering. However, these are often bulky, expensive and difficult to tune. Their selectivity is governed by the percentage frequency separation of the wanted and unwanted signals, selection becoming more difficult as percentage separation decreases. Because of these difficulties, any techniques providing selectivity, such as those illustrated in Figs. 1 and 2, are of potential interest in frequency synthesis.

Figure 3 shows a situation similar to that of Fig. 2(a) but with $f_2 \neq mf_1$, and no forced phase synchronism. The spectral envelope in Fig. 3(a) is again the $\sin x/x$ shape centred on f_2 . However, in this case, components do not in general occur at harmonics of f_1 , but at $f_2 \pm mf_1$. This follows from the fact that the waveform is no longer periodic at f_1 but at some rational fraction of this and the periodicity is no longer controlled only by f_1 but by the combination of f_1 and f_2 .

The process shown in Fig. 3(a) is of course normally referred to simply as "mixing" and it provides a means of adding or subtracting frequencies. As with multiplication, a major problem is the rejection of unwanted frequencies. The situation is normally worse than that shown in Fig. 3(a) because harmonics of f_2 are present at the input or generated in the switching process. Figure 3(b) and (c) illustrate the problem created. In (b) the ratio f_2/f_1 is high, making it difficult to separate $f_2 +$



f_1 from f_2 and $f_2 + 2f_1$. Components from the spectrum centred about the second harmonic of f_2 will also come close to $f_2 + f_1$. However, in this case, they will only be small amplitude signals.

In Fig. 3(c), the ratio f_2/f_1 is low, easing the separation of $f_2 + f_1$ from f_2 . The problem is now that of separating out $2f_2 - f_1$. Filtering problems therefore set upper and lower boundaries on mixing ratios.

The problem in the case shown in Fig. 3(b) can be eased by using the selective characteristics of the spectral envelope. To start with, the use of equal periods will suppress the $f_2 + 2f_1$ component. If, in addition, the off period is replaced by a phase reversed f_2 , we can

achieve cancellation of f_2 as well. This is illustrated in Fig. 4(a), the nearest components are now $2f_1$ away. In practice this cancellation will not be complete; for instance, if the periods are not equal, the result will be as shown in Fig. 4(b). However, very useful attenuations of 20 to 40dB, can be obtained.

These simple considerations of multiplication and addition lead us to some of the practical circuits used for these operations.

Harmonic generation can be achieved by a variety of standard pulse circuits, and modern integrated circuits can generate pulses with harmonics up to about 1GHz. Traditionally, of course, class C amplifiers have been used in transmitters as frequency multipliers

and are normally a harmonic generator with some frequency selectivity incorporated.

Multiplication to very high frequencies, tens of GHz, is possible with step recovery diodes (s.r.ds). The s.r.d. functions as a switch with switching times in the region of 50×10^{-12} seconds. Figure 5 shows a particular arrangement using a s.r.d. with a resonant transmission line. The diode acts as a short circuit while passing forward current, and continues to maintain this condition after the current reverses. When all the current carriers have been swept out of the diode it rapidly switches to an open circuit, shunted by the reverse capacitance of the diode. With a suitable drive level and bias voltage V_b it can be arranged that there is a maximum current in the inductor L at the instant of switching due to L resonating with the diode capacity and the line impedance R_0 , through the harmonic by-pass capacity C . The pulse generated across the diode travels down the line, is reflected at the open circuit, and returns to the diode. At this time the diode switches to forward conduction again. The energy therefore continues to be reflected up and down the line. It can be seen that the output is characterised by a forced phase synchronism of the waveform on the line. The frequency is therefore a harmonic of f_1 and independent of the line tuning. The line tuning of course, provides the selection of the required harmonic.

The quenched oscillator shown in Fig. 6 gives a similar spectrum to that of the step-recovery-diode multiplier, though this is normally used at lower frequencies. Transistor Tr_1 operates as a grounded-collector Hartley oscillator tuned by L and C to approximately the required harmonic frequency. Transistor Tr_2 is switched by f_1 causing the LC circuit to be heavily damped, thus stopping oscillation. The damping period must be sufficient to dissipate the stored energy in L and C . When Tr_2 switches off the circuit transient causes oscillation to restart, in the same phase every cycle of f_1 .

In our initial consideration of mixing, Fig. 3, it was seen that simple switching of one frequency by another produces the $f_2 \pm f_1$ frequencies wanted for addition or subtraction. Mixer circuits normally apply such a switching action, though f_1 and f_2 are usually in sinusoidal form.

Figure 7(a) shows a transistor mixer in simplified form. Source f_2 switches the transistor into conduction on positive half cycles and the lower amplitude source f_1 modulates the amplitude of the pulses producing the output shown. This has components at $f_2 \pm f_1$, but simple inspection shows that there are also components at f_1 and f_2 .

The possibility of cancelling components has already been mentioned and Fig. 7(b) is a reasonably obvious modification of Fig. 7(a) to provide cancellation. In (b) the two transistors

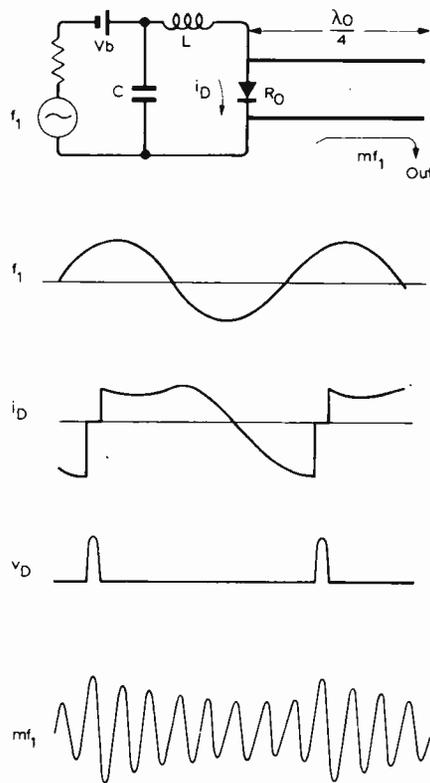


Fig. 5. Circuit diagram and waveforms for a step-recovery-diode multiplier on a resonant transmission line. When conditions are right, the diode switches to open circuit and the pulse generated across it travels down the line, is reflected, and returns to the diode. At this time the diode switches to forward conduction again and the process is repeated - the pulse being reflected up and down the line at a frequency equal to a harmonic of f_1 , depending upon the line tuning. See text.

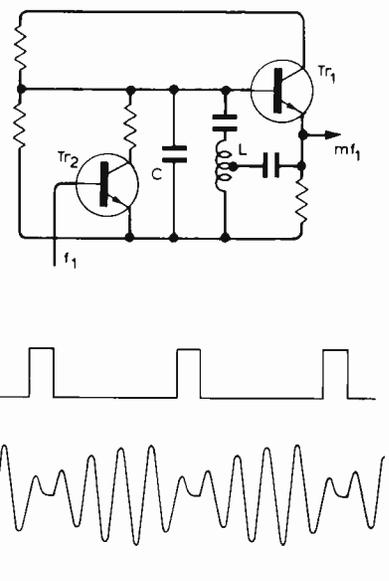
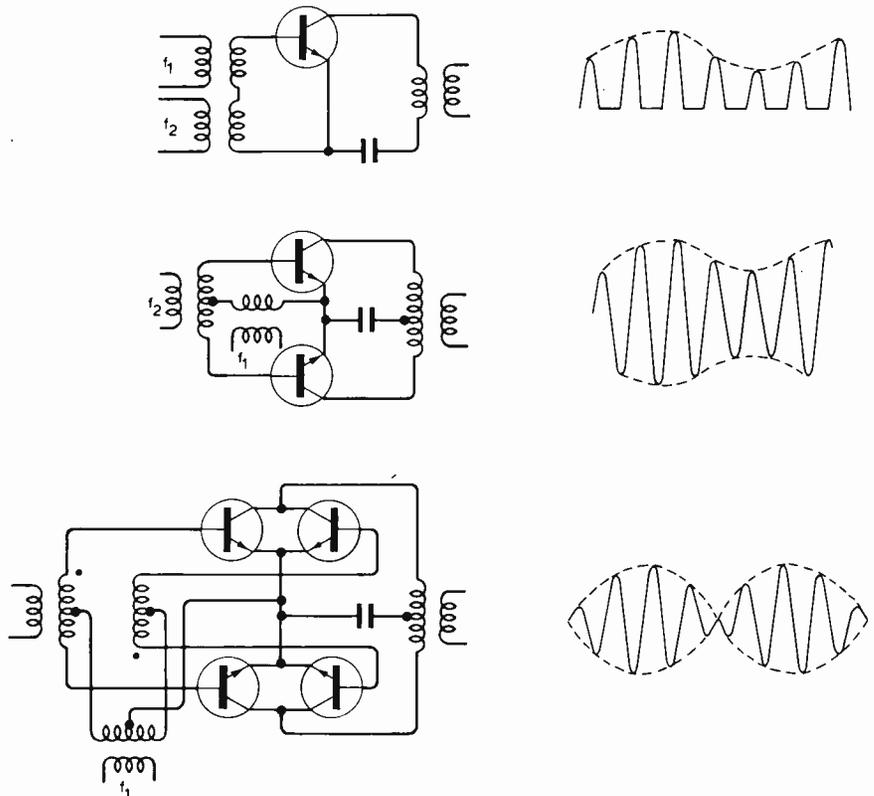
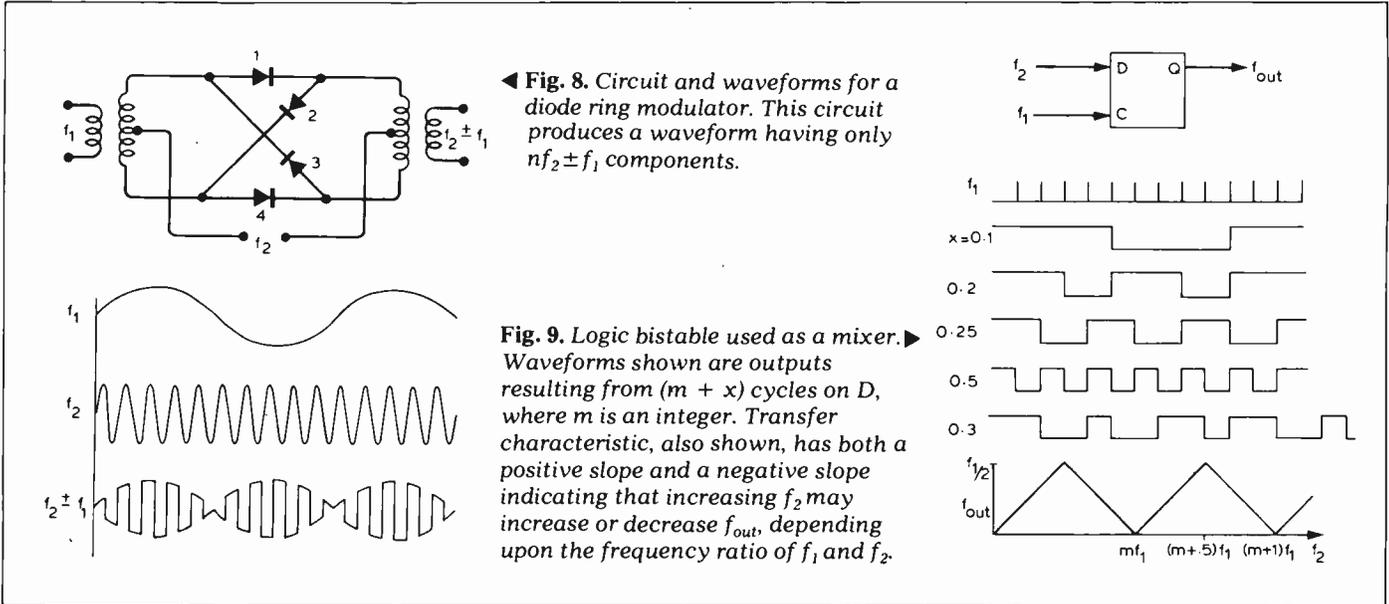


Fig. 6. Quenched oscillator circuit which provides a similar spectrum to the multiplier in Fig. 5. See text.

Fig. 7. Circuit (a) is a simple transistor mixer which produces frequency components at $f_2 \pm f_1$, f_1 and f_2 . In circuit (b) the f_2 components in the two transistors are in antiphase, resulting in a waveform having no f_1 component. Circuit (c) consists of two type (b) circuits and produces a resultant waveform having neither f_2 or f_1 components.





◀ Fig. 8. Circuit and waveforms for a diode ring modulator. This circuit produces a waveform having only $nf_2 \pm f_1$ components.

▶ Fig. 9. Logic bistable used as a mixer. Waveforms shown are outputs resulting from $(m + x)$ cycles on D, where m is an integer. Transfer characteristic, also shown, has both a positive slope and a negative slope indicating that increasing f_2 may increase or decrease f_{out} , depending upon the frequency ratio of f_1 and f_2 .

operate similarly to that in (a) but the f_2 components are in opposite phases in the two transistors. The resulting output waveform has no f_1 component and is in fact the waveform of a conventional amplitude modulated carrier.

The logical step from the method used in Fig. 7(b) is to arrange the cancellation of f_2 as well, as shown in Fig. 7(c). This consists of two circuits of the type shown in (b) with antiphase f_2 signals. The phase of the zero crossings, at the frequency f_2 , reverse every cycle of f_1 , and there is in fact no frequency component at f_2 . The waveform shown is that of a suppressed carrier, double sideband signal.

Mixers that suppress only one of the input frequencies are called balanced and those that suppress both of the input frequencies are called double balanced.

While the transistors in the mixers of Fig. 7 are conducting, they can introduce spurious mixing products due to

the nonlinearities of the transistor characteristics. Many mixers use diodes or transistors as near ideal switches and the most common circuit is the diode ring modulator shown in Fig. 8. In this modulator the amplitude of f_2 is much larger than f_1 and switches on diodes 1 and 4, or 2 and 3. This chops the f_1 signal as shown, producing an output having suppressed f_1 and f_2 components. High drive levels and fast-switching Schottky diodes are used to make the circuit operate as an ideal switching mixer. The only terms in the output are $nf_2 \pm f_1$, giving good separation of higher order products; that is, no $nf_2 \pm mf_1$ terms.

Another type of mixer uses a logic bistable. This 'D' flip flop, which is readily available in integrated circuit form, when clocked, simply transfers the logic state on its D terminal to its output. If two separate frequencies are applied to the D and clock inputs, the output waveform will contain a beat frequency pattern varying as the two

waveforms move with respect to each other. Figure 9 shows the types of output waveform obtained with such a mixer and the mixer transfer characteristic.

Any number of cycles may occur on the D terminal between clocking pulses. If there are an exact integral number of cycles, the logic level on D at the clocking instants will always be the same. The output will be d.c. If there are $m + x$ cycles, where x is a fraction of a cycle, the level at D will vary between some successive clocking instants. Where $1/x$ is an integer, the output will be a simple waveform with a fundamental frequency equal to xf_1 or $(1 - x)f_1$, see Fig. 9.

When $1/x$ is not an integer, the fundamental frequency will be a subharmonic of xf_1 , with xf_1 being present as a harmonic of that.

An important point to note is that there is a positive slope and a negative slope on the mixer transfer characteristic. This means that increasing f_2 may increase or decrease f_{out} , depending on the frequency ratio of f_1 and f_2 .

The author

Raymond Thompson was born in Belfast but has spent most of his life in England. He was educated at Cheltenham and Birmingham Technical Colleges, and in 1960 joined the Plessey Company at West Leigh. In 1966 he went to Westinghouse Research in the USA where he worked on the design of high power converters and inverters, including special inverters for h.f. fluorescent lighting and lightweight power supplies for X-ray units. In 1969 he rejoined Plessey and he is currently at their Roke Manor Establishment designing v.h.f. and u.h.f. digital radio systems for military applications. Raymond Thompson has had several papers published, including two articles in *Wireless World*.



Part 2 of this article will discuss frequency division circuits which use digital binary counters. It will also explain how prescalers can be used to extend the frequency range of such circuits.

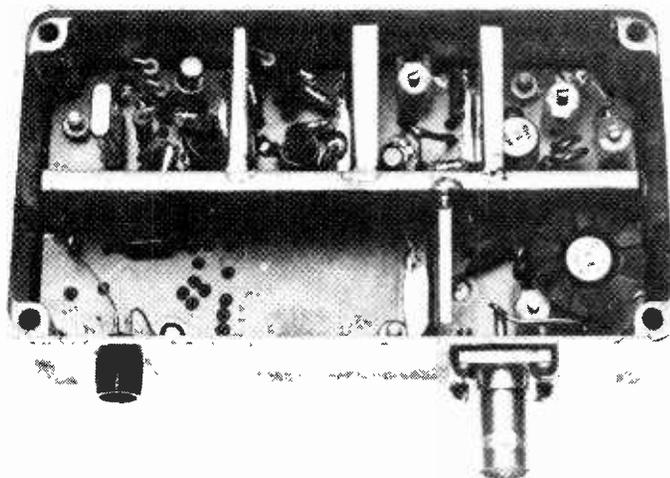
Versatile microwave source

Multiband unit comprises u.h.f. source and step recovery multiplier

by G. D. Lean, B.Sc., A.R.C.S., M.I.E.E.

Designed for simple communication links using Gunn diodes in either professional or amateur equipment, this unit improves frequency stability and reduces bandwidth. It can also be used as a replacement for klystrons in radar assemblies and communication systems. Comparing favourably in noise output to many earlier designs of solid-state sources it offers simplicity and improved reliability.

THIS SIMPLE MULTIPURPOSE microwave source is capable of providing milliwatts of r.f. power at frequencies between 4.5 and 8GHz. It can be used as a local oscillator or as a lower power transmitter or driver with frequency modulation of up to 80kHz peak-to-peak deviation. By using scaled versions of the final



Modulation amplifier Tr_6 of Fig. 1 can be situated in bottom left-hand compartment of this u.h.f. driver.

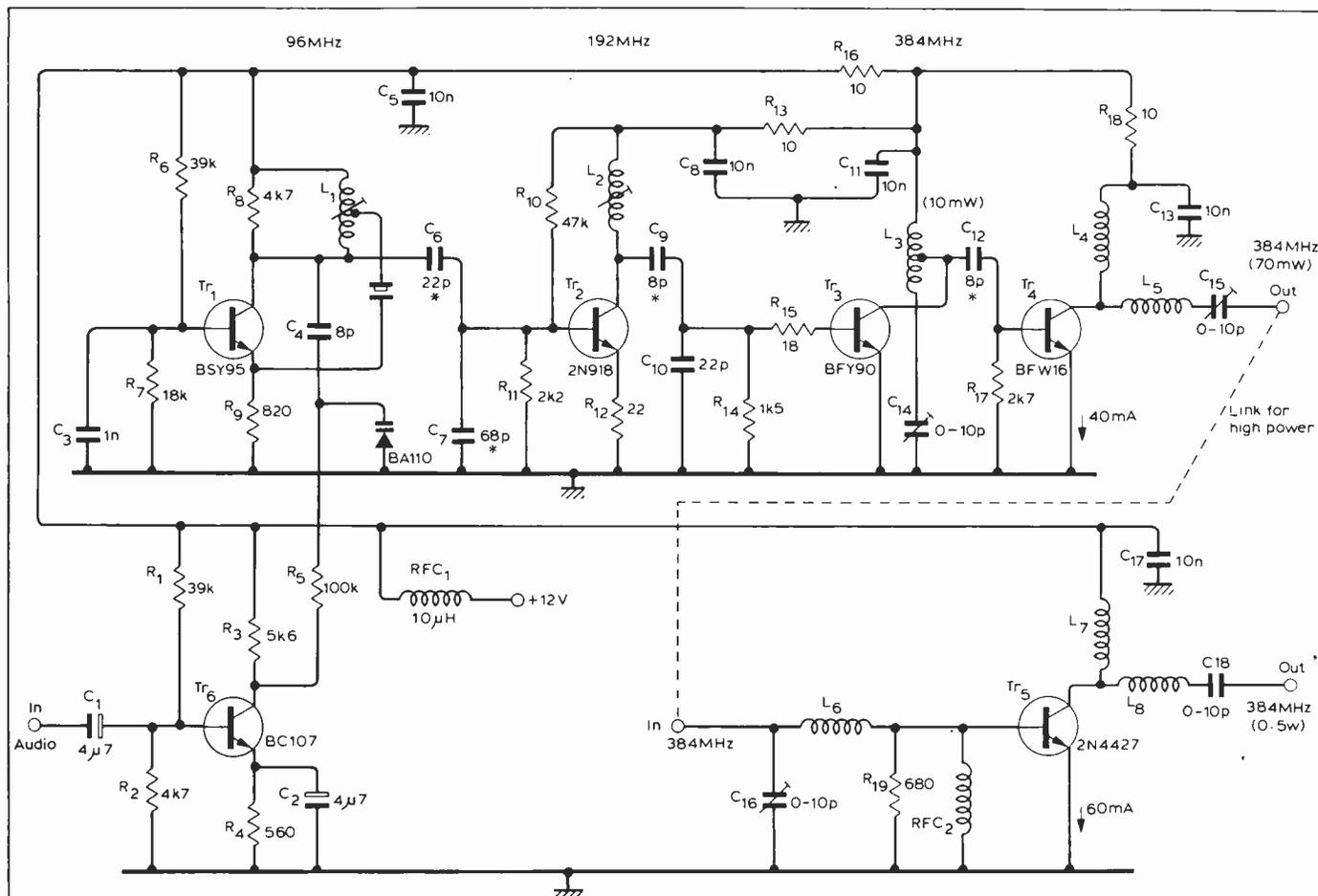


Fig. 1. Capacitor valves marked thus * in this driver circuit should be adjusted on test for best results. Variable types are Mullard C80905/02. Decoupling capacitors are ceramic discs and others have a polystyrene dielectric. See p. 56 for inductor details.

multiplier, frequencies in the range 2.5, 4.5GHz and 8.0 to 12GHz can also be obtained. Output powers in the range 1 to 10mW can be achieved depending on the output frequency and the varactor in use. The source comprises three separate units which can be built and tested separately: a u.h.f. driver, step-recovery multiplier, and harmonic selection filter.

Driver circuit

The driver circuit develops about 500mW of power at around 384MHz, with an optional output of 50mW so that Mullard modules BGY22 can be driven to operate high power multipliers for transmitter or up-converter applications. The circuit shown in Fig. 1 is an adaptation of earlier designs¹ originally based on a compact transmitter design by G3TDZ. It uses readily available cheap transistors and the complete unit can be built for less than £10 including the crystal.

The crystal oscillator stage uses a fifth overtone crystal to give positive feedback from collector to emitter with L_1 tuned to the overtone frequency of 96MHz. Slight pulling of the oscillator can be achieved by adjusting L_1 so that the exact crystal frequency is obtained. The variable-capacitance diode and capacitor C_4 form part of the tuned circuit together with L_1 , C_6 and C_7 , and changes in varicap bias cause slight frequency shifts in the oscillator. The resultant narrow band f.m., with a peak deviation of about 1kHz, at the crystal frequency, produces adequate deviation for telephony communication when multiplied up to the microwave bands. Transistor Tr_6 gives some gain for the modulation voltage and provides d.c. bias for the varicap diode via R_5 . Resistor R_8 damps coil L_1 and prevents any tendency for self-oscillation of Tr_1 due to its internal collector-emitter capacitance. Oscillation should cease if the crystal is removed and should not vary by more than 10kHz as L_1 is tuned. Capacitors C_6 and C_7 form a capacitive tap to match into the base of the first doubler stage. Components L_2 , C_9 and C_{10} form a tuned circuit at 192MHz and provide matching into the second doubler Tr_3 . About 10mW can be measured at L_3 and C_4 (384MHz) by using a one-turn coupling loop feeding a power meter. Resistor R_{15} is a base "stopper" to prevent parasitic oscillations which are common in BFY90 amplifiers.

Matching into the first amplifier is achieved by tapping the coil L_3 and selecting C_{12} . About 70mW can be measured at the output of Tr_4 , which is driven into conduction by the self bias created across R_{17} . Components L_5 , C_{15} , L_6 , C_{16} form a two-stage matching network going through 50 ohms at the link point. For low power requirements the final output can be connected to C_{15} instead of the wire link. For higher power Tr_5 gives the extra gain to provide 0.5-watt output at C_{18} . This final

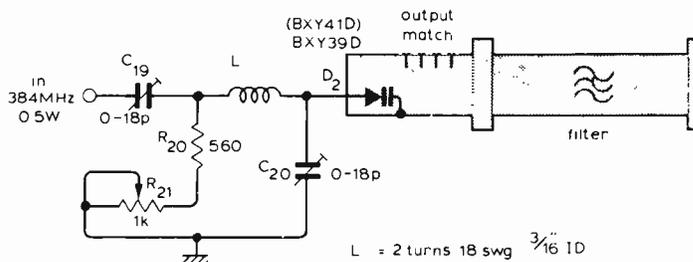


Fig. 2. Step recovery multiplier gives a comb of frequencies spaced at 384MHz, one of which is filtered out, to give a few milliwatts at 5.76GHz. (Use BXY41D for X-band). Variable types are Mullard C80905/03.

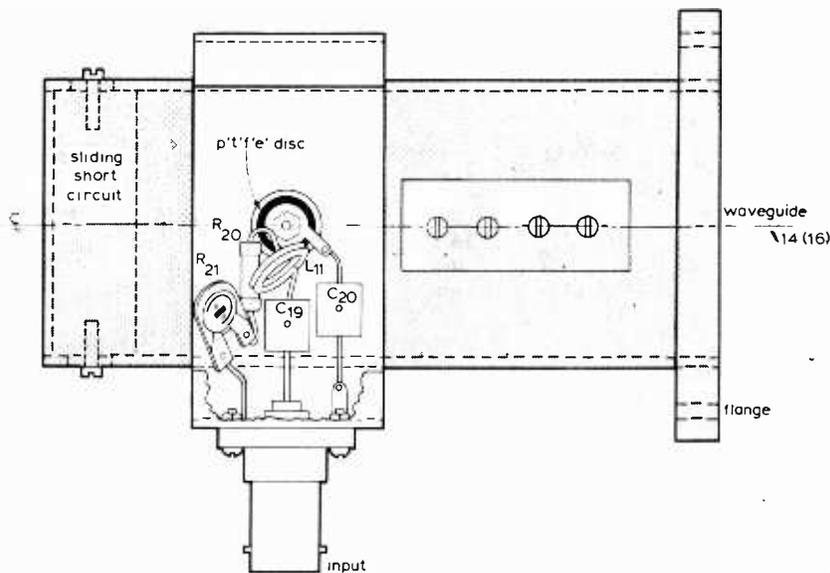
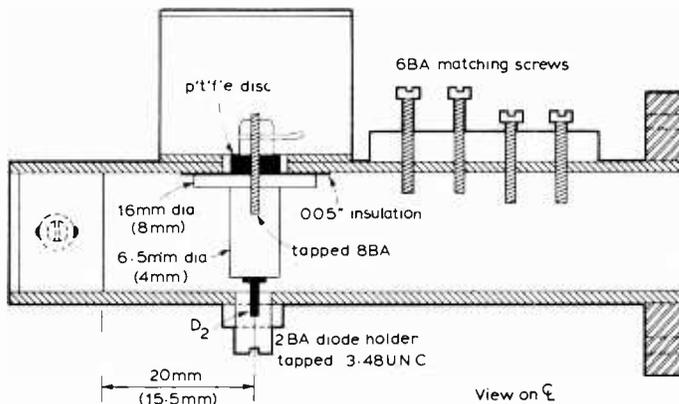
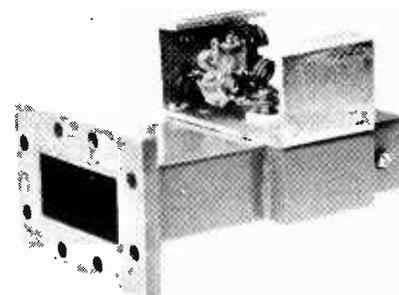


Fig. 3. Step recovery multiplier unit and matching for 5.6 to 6GHz. Dimensions in brackets apply for 10.4GHz in waveguide 16.

stage draws 60mA and is self-biased through RFC₂ and R₁₉.

Driver construction

The unit is constructed on double-sided glass-fibre board, etched on one side as shown in the diagram. For "one-off" production the easiest way is to photocopy the pattern and stick the copy onto a suitably-sized piece of board. Then drill through the paper into the board and clean up the holes carefully. Mark out the circuit using an etch-resist pen or resistant transfers and completely paint over the earth plane side.



This view of the step-recovery multiplying circuit omits screening cover and matching screws.

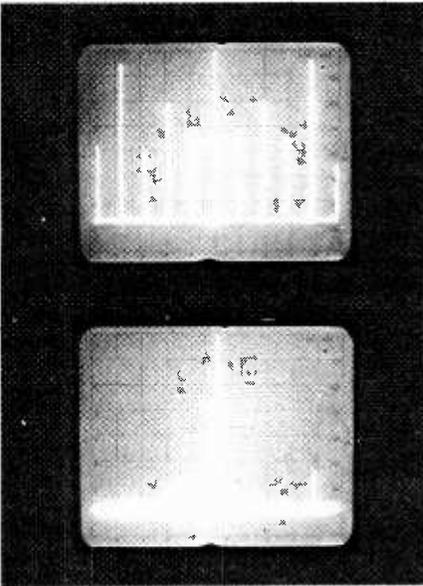


Fig. 4. Multiplier output at 5.76GHz has 300 kHz bandwidth (top), improved to 30kHz by Fig. 5 filter (bottom). Horizontal scale 100MHz/div.

Then etch the board and remove the resist. Countersink the holes on the earth plane side and fit pins to provide the external connectors. Copper or tinplate screens can then be soldered to the top side and coils L_1 and L_2 fitted. Fit the components, starting with the oscillator stage and test each stage before starting the next. Finally when the board is finished and tested mount it in a small die-cast box on 6BA nut spacers beneath the underside of the board. Modulation input, power supply and r.f. output connections can then be taken through convenient sockets in the box sides.

Multiplier design and construction

The multiplier shown in Figs. 2 & 3 is a C-band version of an X-band design² by P. Tunbridge (G8DEK) which uses a Mullard varactor diode to generate a comb of frequencies, one of which is selected by the output filter. The input matching components are L_9 and C_{19} while C_{20} provides some capacitance trimming to the input capacitor formed inside the waveguide by the shaped diode-support pillar. Resistors R_{21} and R_{20} provide a d.c. bias return for the varactor; these are low impedance to give improved high-order multiplication. Matching into the waveguide is provided by four tuning screws in the broad face of the waveguide together with a sliding short circuit which provides a resonant cavity at the output frequency.

First fabricate two cover mounting blocks, a diode holder nut and a matching screw block from brass, and solder these to the waveguide, together with a suitable flange. The matching screw holes can be drilled and tapped into the block and guide and only three holes spaced at $\lambda/8$ are all that are strictly necessary. However, if four or five holes are drilled at about 7mm spacing, a wider range of output

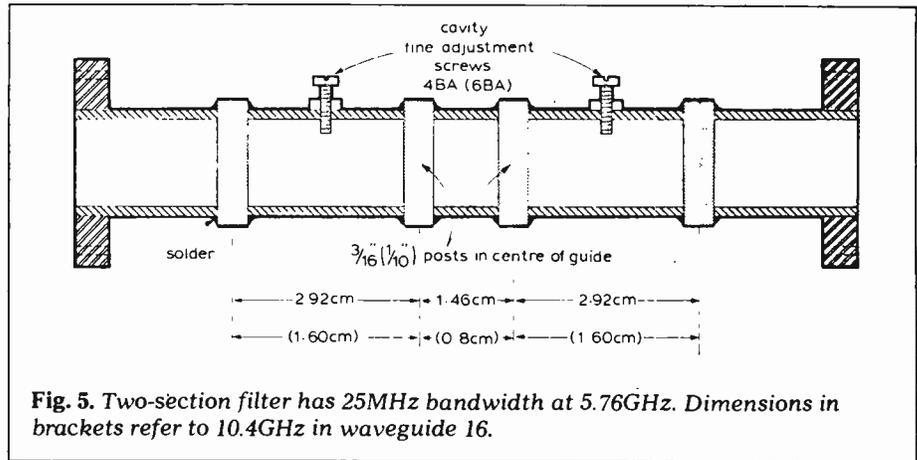


Fig. 5. Two-section filter has 25MHz bandwidth at 5.76GHz. Dimensions in brackets refer to 10.4GHz in waveguide 16.

frequencies can be accommodated. Two slots for the short circuit can be cut in the narrow face of the waveguide and a 3/16-in hole in the broad face for the connection to the diode support pillar. Underneath the guide the diode-holder nut should be cleared with a 2BA tap which should also continue the thread through the waveguide wall. Screwing a pointed tap right the way through until it touches the top wall will mark the position for the connection hole which can then be pilot drilled through the nut. A diode support pillar can either be turned down from a brass bar or made up of a disc of 16 s.w.g. brass or copper soldered to some 1/4-in brass rod. The broad end is tapped 8BA for the connection screw. A disc of insulation material such as Micalex or just Sellotape is placed on the top of the pillar which can then be slid into the waveguide under the 3/16in connection hole. A thin 3/16-in dia insulating washer is then dropped in the hole to centralize the pillar. A further 1/4-in insulating washer is used with an 8BA bolt and tag to clamp the pillar and provide electrical connection to the diode.

The diode holder is made of 2BA copper studding with a tapped hole in one end for the diode. Some diodes have untapped ends and a 1/16-in hole is all that is required to mount them in the holder. Two side cheeks of 16 s.w.g. aluminium are bolted to the cover-mounting blocks with one supporting the BNC input socket. The rest of the input components can then be soldered into circuit.

Testing

Select a Mullard BXY39D for 5.7GHz (a BXY40D may give more output around

6GHz). This should be screwed into the diode holder using a dab of thermal grease on the threads. The holder is then screwed into the multiplier unit so that the diode flat contacts the bottom of the support pillar firmly, but not too tightly. A lock-nut should then be tightened on the diode holder.

With the diode in place the v.h.f. source can be connected and powered. A 96MHz crystal will result in an output of the driver source at 384MHz, which when fed into the multiplier produces a comb of frequencies from about 4 to 10GHz with a spacing of 384MHz. Some products of 96MHz and 188MHz are also present, depending on the purity of the driver output.

Alignment of the multiplier is best carried out with a spectrum analyser, but if the filter described below is made it is possible to align the multiplier with only a diode detector or power meter. Using the analyser the required final frequency is displayed together with the two adjacent 384MHz spaced frequencies. Frequencies every 96MHz will also be present about 30dB below the 384MHz harmonics. First slide the short circuit in or out until maximum power is obtained at the output frequency. Then adjust C_{19} , C_{20} and R_{21} again for maximum power of the output. There is some interaction between R_{21} and the capacitors; R_{21} should be set in several positions whilst C_{19} and C_{20} are adjusted for optimum each time. When no more improvement in output can be obtained from the input tuning, a 6BA tuning screw can be tried in each hole in turn until the best output is obtained. It is then left in the hole. More screws should be tried in the holes until the required output frequency is peaked about 6dB more than other harmonics with the ± 96 MHz products well down, as shown in Fig. 4 (top). Output purity can be much improved with the filter, see Fig. 4 (bottom).

Filter

The filter is a two cavity design, adjustable from about 6.1 to 5.2GHz with the dimensions given. The design was achieved by accident as a gross mathematical error resulted in incorrect theoretical dimensions which work

Inductor details for Fig. 1

- L_1 five turns 22 s.w.g. wire tapped one turn from "cold" on 3/16-in dia former and slug.
- L_2 three turns 22 s.w.g. wire on 3/16-in dia former and slug.
- L_3 half turn 18 s.w.g. wire loop tapped halfway.
- L_4, L_5 one turn 18 s.w.g. wire.
- L_6 1/2-in of track.
- L_7, L_8 one turn 18 s.w.g. wire.
- RFC₁ 10 μ H.
- RFC₂ two turns 26 s.w.g. on ferrite bead.

well in practice. Basically if a post is put in the centre of waveguide it creates a large susceptance which reflects the incident wave. If another post is a quarter wavelength away from the first it will create an equal and opposite susceptance which will cancel the effect of the first at one frequency. Two such pairs of posts spaced at a quarter wavelength will cancel any residual susceptance giving unhindered matched transmission to the design frequency. This simple theory doesn't work exactly in practice and the modified dimensions should be adhered to. The dimensions can be scaled for other frequencies, remembering to scale waveguide dimensions to λ_g rather than the free-space wavelengths. Scaled versions have been made up to 11GHz and perform just as the 5.7GHz version shown in Fig. 5.

Construction is fairly obvious and for quick prototypes the posts need not be soldered into the guide but just pushed into tight fitting holes drilled through the guide. As the 4BA tuning screws are screwed in, the pass-band frequency is reduced and the response at mid-adjustment is shown in Fig. 6. Without a spectrum analyser the filter can be adjusted first on a fundamental signal source such as a Gunn diode³ or klystron whose frequency has been adjusted using a wavemeter. Once the filter is aligned on the correct frequency the multiplier can be adjusted for maximum output through the filter. The filter can be slightly readjusted for maximum

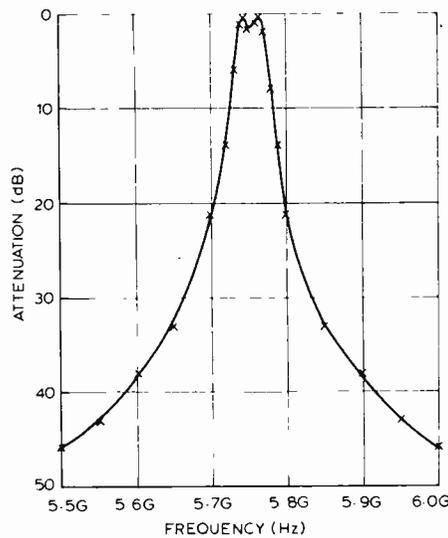


Fig. 6. Response of two-cavity filter as measured at mid-adjustment of the tuning screws

output on the exact frequency multiple; it is unlikely that the wrong harmonic has been chosen unless the original frequency measurement using the wavemeter was more than 188MHz out of true. Provided an output of 3 to 5mW is obtained through the filter it is unlikely that any spurious products are present.

A single cavity version of the filter can be made for simple equipment when total suppression of other products is

unnecessary. The suppression of out of band products is about half that of the larger two-cavity design.

The multiplier input will accept frequencies in the range 350 to 450MHz and so 432 MHz low-power f.m. transmitters can feed into the multiplier via an attenuator, although for amateur band use not all the bands can be covered as would be possible with a single 384MHz generator. It is also possible to change crystals in the generator to produce local oscillator frequencies suitable for use with low intermediate frequencies. But to obtain the best results high intermediate frequencies (140MHz) should be used to reduce local oscillator noise getting into the receiver. This does mean that a separate local oscillator is required for each band, but for a low-power transmitter a single modulated u.h.f. source can be used with several multipliers for all the bands.

A printed board pattern and component layout appear on page 63.

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2. Tonbridge, P. M. Multiplier 10GHz varactor, *Radio Communication* 1976, page 202.
3. Hosking, M. W. Microwave voice link, *Wireless World*, vol. 83, October 1977, pages 49-52, and November 1977, pages 69-71 & 92. □

continued from page 40

them situated peripherally and at present served with Radio 4 by local m.w. booster stations will get a much weaker signal and one more vulnerable to electrical machine noises:

But what about the discriminating public; i.e., the people who appreciate the merits of the f.m. service and are prepared to pay for it? They have already found that many BBC programmes are not available on v.h.f. It is all very well to be repeatedly reminded that with three-waveband receivers they can get all that the BBC offers, but these reminders studiously refrain from mentioning that owing to the increasing practice of subdivision many of the programmes are, and will continue to be, available only on a second-class service, i.e.,

- (1) No stereo
- (2) Lo-fi (restricted audio frequencies)
- (3) Greater liability to noise

As regards those programmes that are on v.h.f., in fairness to Mr Redmond again I must recall that his impatient awaiting was for portable v.h.f. radios. Ever since v.h.f. receivers became available in quantity, some of the better-class non-portable models have

included instant tuning. The wonder to me is that anyone has the nerve to market any such model without this facility. And I cannot see any major problem in extending it to portables. In fact, being no longer in touch with the trade I was really astonished by the authoritative statement that such sets were still being awaited — hence my present eruption into these now usually erudite and well-behaved pages.

Having, as I said, every sympathy with the general public in rejecting as totally outmoded and unacceptable a technology that requires from them a skilled and time-wasting procedure every time they want to change programmes, I suspect that there are many households that have several sets, each permanently tuned to one of the limited number of channels they use. Perhaps this wasteful solution is so good for the radio trade as to account for the tardiness of that trade in fulfilling Mr Redmond's earnest hopes.

Finally I will seize this opportunity to protest about what I regard as that design monstrosity advocated — I am sure very reluctantly and unwillingly — by the BBC; the combined f.m. and a.m. receiver. The design requirements are

so different that a receiver which works on both is almost two different sets in one. A few years ago an even worse monstrosity was put before the British public: the combined 405 and 625 line TV set, which consisted virtually of two quite different — opposite, really — receivers with a very complicated multi-contact (and therefore doubtfully reliable) hard-to-turn switch. I refused utterly to admit such a thing to my home, and did without colour until all-625-line sets were available. The solution to that problem was to provide all the programmes on one system.

The same solution is the only right one for radio. So my impatient waiting is for a broadcasting system in which all the programmes are obtainable on v.h.f. If they, or some of them, are also available on a.m., so much the better. But we ought to be able to go back, if we wish, to the v.h.f.-only receiver. It can be done by making available here more of the 88 to 108 MHz v.h.f. broadcasting band for the purpose of broadcasting, thus allowing the industrious student, the hard-working housewife, the parliamentary enthusiasts (if any), the pop addict and the music lover each to enjoy the benefits of f.m. stereo undisturbed.

Tunable audio equalizer

Flexible parametric equalizer with variable Q

by Martin Thomas

MOST AUDIO EQUALIZER circuits represent a compromise between cost, facilities and ease of use, and the Baxandall tone control has been by far the most successful design. For domestic audio equipment its simplicity and ease of use outweigh its disadvantage of providing only a limited degree of equalization, although the circuit can be modified to increase its flexibility¹. Clearly, however, the "bass" and "treble" subdivision of the audio band is insufficient for many purposes, and the graphic equalizer approach of having a larger number of frequency bands becomes necessary. The only problem with this approach is that a large number of controls must be used to cover the audio band if the individual frequency bands are narrow, so even with this circuit the number of controls is a compromise.

Unquestionably the most versatile equalizer is the parametric, or tunable, type. In its simplest form it can consist of only a single boost/cut element, but its centre frequency can be varied continuously over a wide range (possibly over the whole audio band), and the Q can also be varied so that either a broad or a narrow frequency band can be equalized. This approach allows almost any equalization requirement to be met with only a small number of such elements, and since the elements are iden-

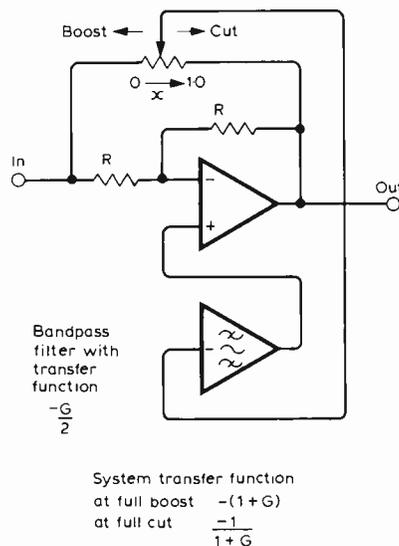


Fig. 1. Basic equalizer design shows how to achieve either boost or cut with a single active element.

tical, no more than are actually needed can be connected together for any particular application. A parametric equalizer may not be so straightforward to use as a graphic equalizer, but once you become accustomed to the rather different controls it's much easier than you might expect.

Circuits of this type have been around

for a number of years, but relatively little has been published about them. In this article I shall take the opportunity to discuss some aspects of the design theory, in addition to describing my own design. The circuit has continuously and independently variable centre frequency, boost/cut amplitude and Q, and also allows a choice of two different sets of boost/cut amplitude-frequency response curves as the control setting is varied; more about that later.

The circuit for a tunable equalizer can be broken into two sections, by which point the basic design is almost complete! The first problem is how to use a single active element either to boost or to cut a given frequency range, and this can be achieved by the circuit shown in Fig. 1. The filter used in the present design is phase-inverting, so its output is connected to the non-inverting input of the amplifier to give overall negative feedback. With this connection there is a gain of two from the filter output to the amplifier output, so the filter transfer function is specified as $-G/2$ to express the system transfer function in its simplest form. When the boost/cut potentiometer is at either end of its travel, the filter is entirely in either the forward or feedback signal path, giving transfer functions of $-(1+G)$ and $-1/(1+G)$ respectively. An exact expression for the transfer function at other control settings will be developed later, but for now notice that when the control is at its midpoint, the forward and feedback contributions will be equal, giving a transfer function of -1 ("flat"). An extension of this circuit to include several filters and potentiometers yields the basic design for a graphic equalizer, of course, and a typical circuit is described in ref. 2.

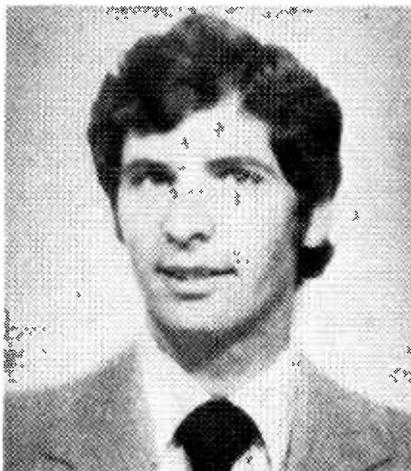
The second problem is the design of the tunable filter. In theory this is very simple, but there are several practical difficulties. Although capacitors can be switched to change the frequency range, the variable control clearly has to be resistive, which rules out some otherwise very promising circuits such as the multi-feedback filter³, since the Q will then also vary. The Wien-bridge configuration does meet this requirement if the resistors in the forward and feedback arms of the bridge are varied together³, but the Q is sensitive to mismatch between these resistors. As the

Three years ago, Martin Thomas left Cambridge University, having collected the B.A. and M.A. degrees in Natural Sciences and a Ph.D. in neurophysiology, to become first a research fellow and later an assistant professor at Boston University. Now, he's returning to the UK to join the Physiology Department at Oxford University.

His audio interests developed while he was at Cambridge, and he designed the prototype for the equalizer there. He says his research activities aren't directly related to his audio and electronics interests, although in practice there's a lot of overlap between them. Using ion-specific dyes to follow changes in ion concentrations inside nerve and muscle cells during excitation is basically a technical problem. "I had to build a sensitive microspectrophotometer to be able to resolve the very small changes in dye absorbance, and it involved a fair amount of electronics."

Would he ever consider moving out of research and into industry? "That depends. Most companies really aren't very interested

in my type of background, and I'd probably have to set up a business of my own. But I certainly wouldn't rule out doing that at some time in the future."



sensitivity increases with Q, the circuit is suitable for use only at low Q, and the long-term reliability is questionable, once the resistor tracks start getting dirty!

The state-variable filter, which is synthesized from integrators, meets the

requirements very well, and has the additional advantage that it is inherently stable even at high Q. Its only drawback is that it uses three operational amplifiers rather than one, but the number of passive components is almost the same as for other circuits,

and it has been chosen for the present design. Fig. 2 shows the circuit diagram. Further information on state-variable filters is given in the appendix and in ref. 4, but the basic equations are reproduced below. Referring to the component values in Fig. 2, the transfer function is

$$\frac{V_o(s)}{V_i(s)} = -\frac{R_2(R_3 + R_4)}{(R_1 + R_2)R_4} \times$$

$$\frac{R_6 C_2 s}{R_5 C_1 R_6 C_2 s^2 + [R_1(R_3 + R_4) / (R_1 + R_2)R_4] R_6 C_2 s + R_3 / R_4}$$

and the bandpass centre frequency is

$$\omega_o = \frac{R_3}{R_5 C_1 R_6 C_2 R_4}$$

For $R_3 = R_4$, and $\omega_o = 1/R_5 C_1 = 1/R_6 C_2$, the transfer function becomes

$$\frac{V_o(s)}{V_i(s)} = -\frac{2R_2}{R_1 + R_2} \times$$

$$\frac{s/\omega_o}{s^2 \omega_o^2 + [2R_1 / (R_1 + R_2)] s/\omega_o + 1}$$

$$= -\frac{2R_2}{R_1 + R_2} \times$$

$$\frac{\omega_o s}{s^2 + [2R_1 / (R_1 + R_2)] \omega_o s + \omega_o^2}$$

Comparison of this last equation with the generalized second-order bandpass transfer function

$$\frac{V_o(s)}{V_i(s)} = \frac{\omega_o A_o s / Q}{s^2 + \omega_o s / Q + \omega_o^2}$$

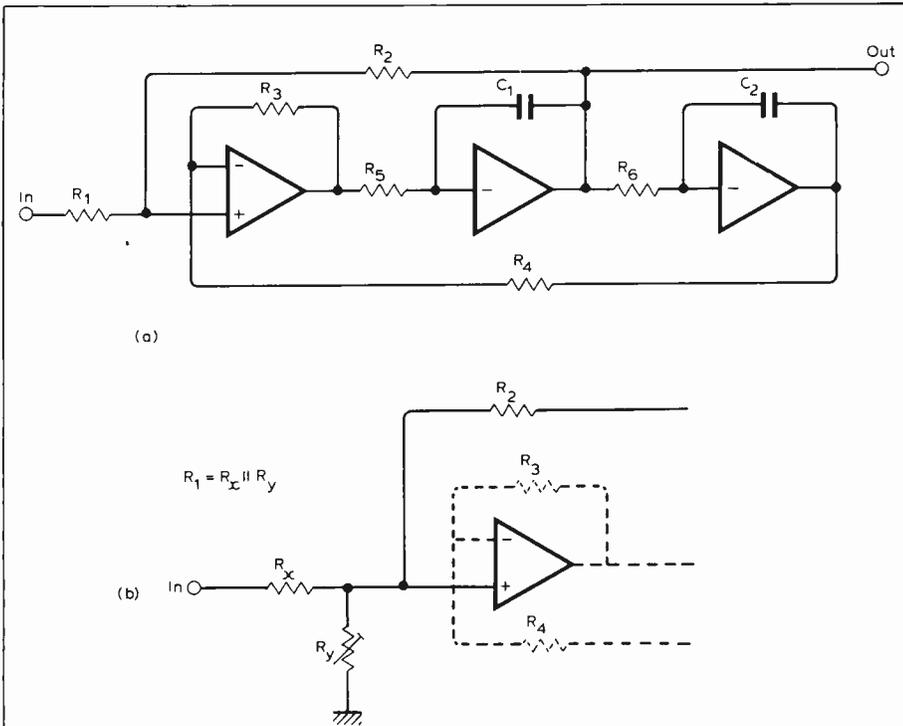
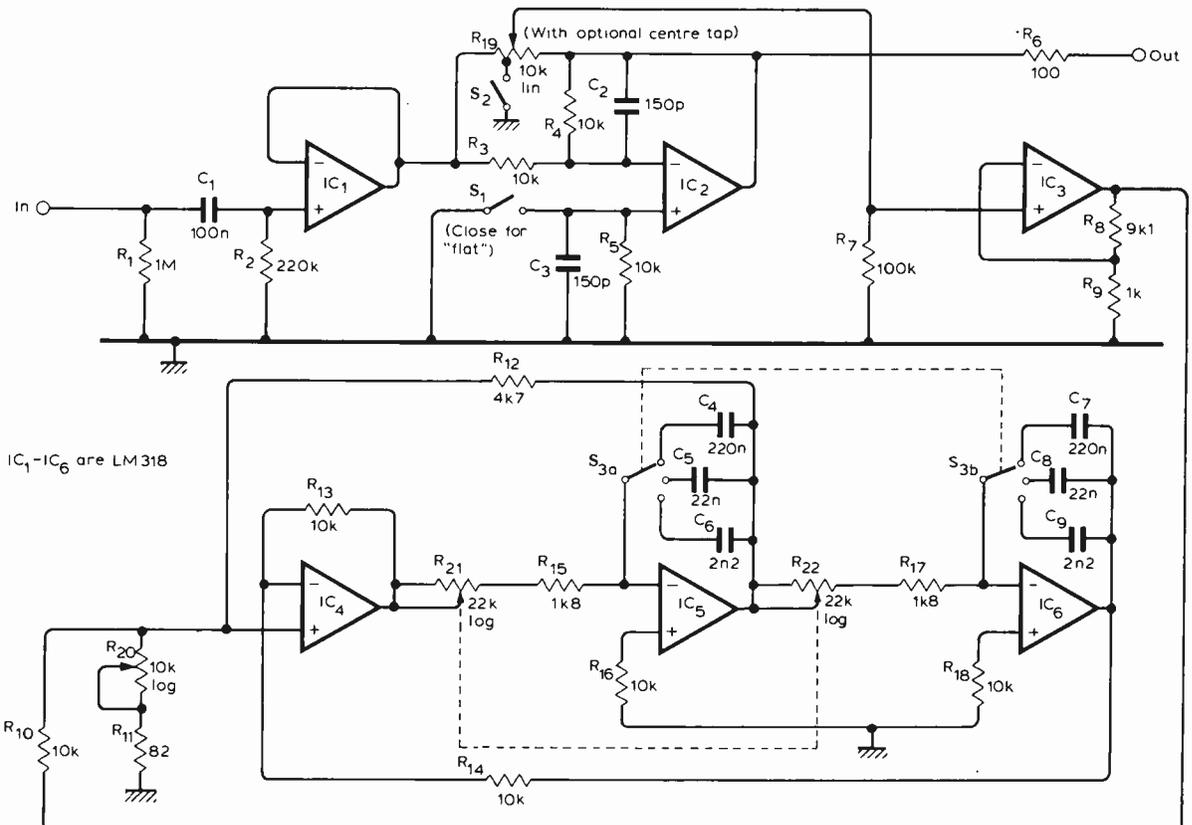


Fig. 2. "State-variable" bandpass filter is inherently stable even at high Q, (a). Modification for constant centre-frequency amplitude, (b). Varying Ry gives independent control of Q.

Fig. 3. Circuit diagram for a single-section tunable equalizer. Ganged resistors R_{21}, R_{22} determine centre frequency, together with range switch S_3 . Boost/cut control is R_{19} while Q is varied with R_{20} . Fig. 5 illustrates function of S_2 .



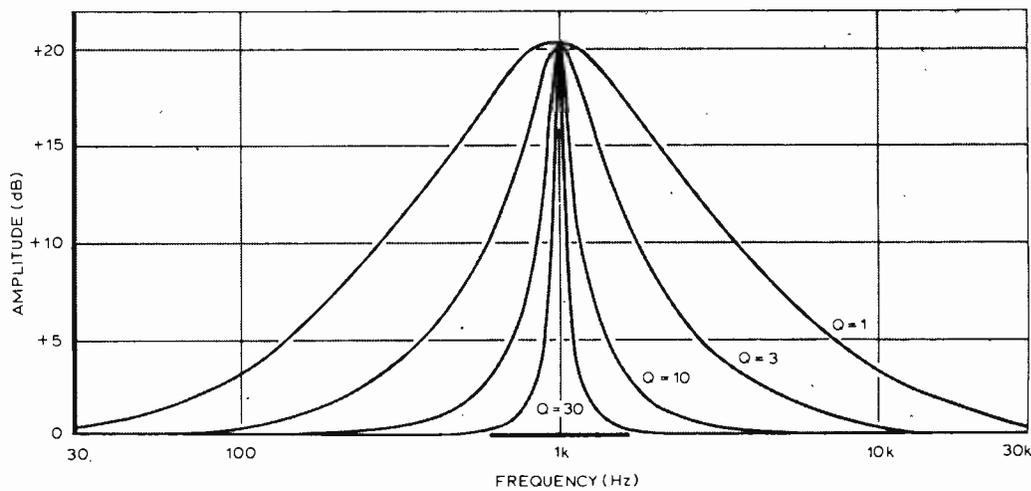


Fig. 4. Effect of varying Q with boost/cut control at maximum boost.

shows that Q is $(R_1 + R_2) / 2R_1$, and the centre-frequency gain A_o is $-R_2/R_1$. By varying R_5C_1 and R_6C_2 together, it is thus possible to vary ω_o independently of Q and A_o . The Q will change if there is any mismatch between these two time constants (although A_o will remain constant), but you can see from the transfer function that the sensitivity does not increase with Q , and hence accurate component matching is not necessary.

The Q can be varied independently of ω_o by varying R_1 or R_2 , but this will also alter A_o , and the relation between A_o and Q is non-linear. Fortunately, a simple modification to the basic circuit, as shown in Fig. 2(b), overcomes these problems. Resistance R_1 is replaced by two resistors, R_x and R_y , so A_o is now $-R_2(R_x + R_y) / R_xR_y$. Resistors R_x and R_y also form an attenuator for the input signal, the gain being $R_y/(R_x + R_y)$. The overall filter gain is the product of these two terms, i.e. $-R_2/R_x$, hence by varying R_y the Q can be varied independently of the centre-frequency gain. The only disadvantage of this modification is that the overall centre-frequency gain of the filter with the component values used in the final design is only 0.5, so additional amplification in the filter path is necessary. The extra amplifier is placed before the filter, where it also provides the necessary low-impedance source.

The complete circuit is shown in Fig. 3. A low-impedance source is provided by IC_1 for the boost/cut control and its associated amplifier IC_2 . The input amplifier for the filter is IC_3 , and IC_4 - IC_6 comprise the filter itself. Although the circuit may appear elaborate, the number of passive components is relatively small, and the size and component count can be reduced further if dual or quad ICs are used. It should be borne in mind, however, that the circuit was designed around the LM318, which is a high bandwidth device. Use of other ICs will result in inferior performance at high frequencies, although it may still be satisfactory for many applications,

and possible substitutions will be discussed later on.

The present design is a general-purpose one, but the range of centre frequencies, boost/cut and Q can easily be modified if required. The centre frequency is determined by the ganged variable resistors R_{21} , R_{22} and by capacitors C_4 to C_9 (see Fig. 3). The variable frequency range is just over tenfold, and capacitor switching gives a total range of three decades, the nominal frequency ranges being 30 to 300, 300 to 3,000 and 3,000 to 30,000 Hz. The range switch S_3 should be a make-before-break (shorting) type, so that the capacitive feedback paths around IC_5 and IC_6 are not interrupted during the instant of switching, otherwise the circuit could oscillate. The sliders of R_{21} & R_{22} should be connected to the clockwise end of their track, so that there is a d.c. path through the control even if the slider loses contact with the track, as the control has to provide a path for the input bias currents of IC_5 and IC_6 . A logarithmic control has been specified, although it will have to be turned anticlockwise to increase the frequency, whereas a clockwise law would be preferable. A clockwise law can be obtained by using a dual antilog control (in which case the sliders should be connected to the anticlockwise ends of the tracks), but this component may not be readily available from most suppliers.

The boost/cut range is determined by the gain of IC_3 and the attenuator at the input of IC_4 . There is a gain of two from the output of the filter network (at IC_5) to the output of the circuit (at IC_2), and a two-fold attenuation at the centre frequency through the filter itself, so the overall gain through the filter path at the centre frequency is given by the gain of IC_3 , which is 10 with the component values shown in Fig. 3. Reference to Fig. 1 shows that the maximum boost and cut are $-(1+10)$ and $-1/(1+10)$, i.e. \pm just over 20 dB, but these values could easily be modified by changing the gain of IC_3 , which is of course $(R_8 + R_9)/R_9$. Input

bias current for IC_3 is provided by the boost/cut control R_{19} and by R_7 , which provides an independent path in case of poor slider contact in R_{19} .

The method of Q variation is as already described with reference to Fig. 2(b), and R_y is represented by R_{20} and its series resistor R_{11} in Fig. 3. The Q range of the circuit is 1 to 30 with the component values shown, and although the upper limit is unnecessarily high for many applications, the value of 30 was chosen simply because it can be achieved over the entire audio range when LM318s are used. It can be reduced by increasing R_{11} . Once again we have a law problem with the control; if R_{20} is a logarithmic control, the Q will increase as the control is turned anticlockwise. An antilog control can again be used to obtain a clockwise law (slider now connected to the anticlockwise end of the track), or a range of fixed, switch-selected Q values can be obtained by replacing R_{20} and R_{11} by a series of fixed resistors, Q being

$$\frac{R_{12}}{R_{10} / (R_{20} + R_{11})}$$

Obviously it is useful to be able to switch out the filtering, and this is achieved by S_1 , which simply shorts the non-inverting input of IC_2 . If the d.c. output of the IC_5 is not exactly zero, the d.c. output of IC_2 will shift when S_1 is closed, but the shift on the prototypes was only a few millivolts and did not cause any audible effects. In fact, the circuit is d.c. coupled throughout (apart from the input to IC_1), and it may be advisable to add a coupling capacitor at the output of IC_2 if the possibility of output offset cannot be tolerated.

The output resistor R_6 is not for protection, but rather to isolate any capacitive load from IC_2 to ensure stability. Capacitors C_2 and C_3 also help to ensure stability by rolling off the amplitude response above 100 kHz. The only other stability precaution — but which is perhaps the most important — is to decouple the $\pm 15V$ supplies with

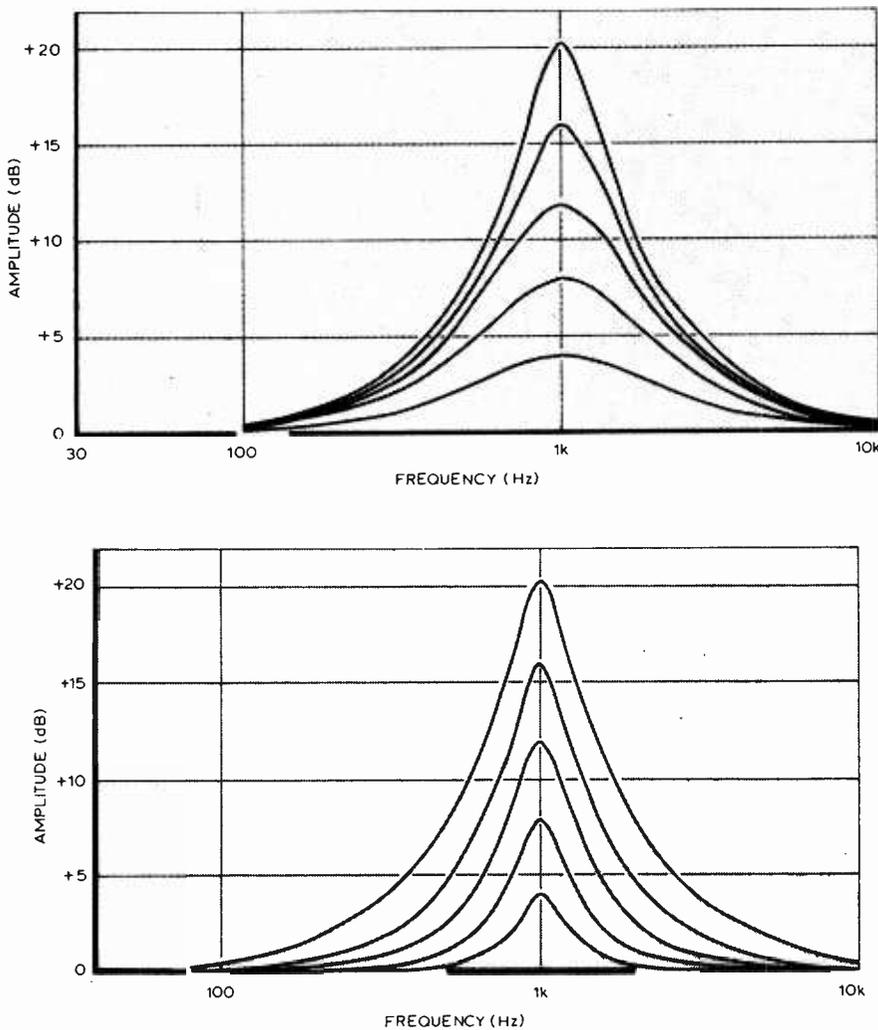


Fig. 5. Two sets of amplitude-frequency response curves can be generated by the circuit with S_2 open circuit (a) and with S_2 closed (b). Corresponding cut curves are symmetrical about log. frequency axis.

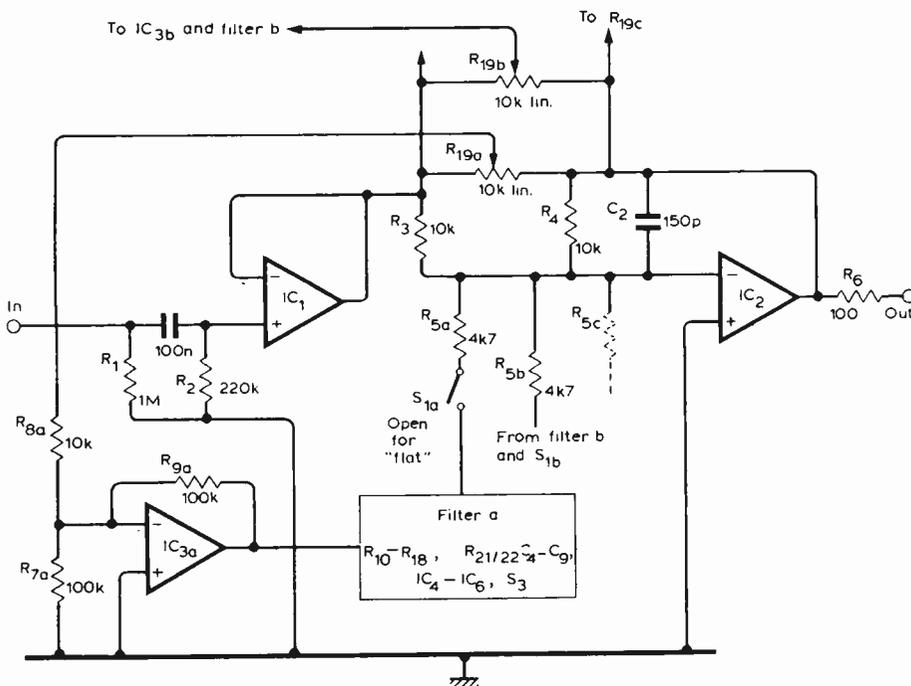


Fig. 6. Modified circuit suitable for use with any number of filter sections.

0.1 μ F capacitors. Such decoupling is very important with high bandwidth devices like the LM318, and if any instability problems are experienced, they can almost certainly be traced to this cause. On the prototypes it was found helpful to connect a capacitor directly between the supply lines in addition to the normal practice of decoupling between each supply line and ground, and the capacitors should of course be sited as close as possible to the ICs. In spite of the impression which may have been gained from these remarks, I was pleasantly surprised by the high stability of the prototypes, and there is no reason to suppose that such results cannot be obtained consistently if standard layout practices are followed.

Performance details of the circuit are given in Figs 4 and 5. These graphs have all been obtained for a centre frequency $\omega/2\pi$ of 1kHz, but the performance for any other frequency in the audio band can be obtained by appropriately shifting the log. frequency axis. Only the boost curves are shown; the corresponding cut curves are symmetrical about the log. frequency axis. Fig. 4 shows the effect of varying the Q control when the boost/cut control is at maximum boost, and gives an idea of the very wide range of equalization curves which can be generated by the circuit. The centre-frequency gain remains independent of the Q control setting as the boost/cut control is varied, but the effect of the boost/cut control on the frequency response is not as straightforward as might be imagined. Fig. 5(a) shows the effect of the boost/cut control when R_{20} is set for a Q of 3, from which it can be seen that the Q is reduced as the control is rotated toward its centre ("flat") position. By a simple modification to the circuit, however, it is possible to generate the curves shown in Fig. 5(b), where the shape of the response remains relatively constant as the boost/cut control is varied. The reason why these two families of curves can be obtained may not be intuitively obvious, but it can be explained by the following analysis.

It has already been shown that the system transfer functions at full boost and full cut are given by $-(1+G)$ and $-1/(1+G)$. At intermediate positions of the boost/cut control R_{19} , both forward (boost) and feedback (cut) signals will pass through the filter. Let the fractional rotation of R_{19} be represented by x , such that at full boost $x=0$ and at full cut $x=1$ (see Fig. 1). Resistor R_{19} will act as a potential divider for the two signals, so the forward signal contribution to the transfer function will be $-(1+(1-x)G)$, and the feedback contribution will be $-1/(1+xG)$, which yields the system transfer function $-(1+(1-x)G)/(1+xG)$. This reduces to the forms previously given for $x=0$ and $x=1$, and to -1 (flat) when $x=0.5$. Gain G can be written as AN/D , where N and D are the numerator and denominator terms of G , and A is the centre-

frequency gain through the entire filter pathway (including the A_0 term, defined previously), which is equal to 10 in the present circuit. The system transfer function now becomes

$$-(D + (1-x)AN)/(D + xAN).$$

Setting ω to 1 for convenience, we have $N = s/Q$ and $D = s^2 + s/Q + 1$.

We are now in a position to explain the curves in Fig. 5(a). When R_{19} is close to the full boost setting, x is close to 0. As x increases (R_{19} rotated away from full boost), the numerator of the transfer function is reduced, but since A is large this reduction will be small compared to the increase in the denominator. Thus when x is close to 0 the transfer function can be approximated by $-AN/(D + xAN)$, which is to say that as R_{19} is rotated away from the full boost position, the change in frequency response can be accounted for primarily by a change in the pole positions. The denominator of the transfer function is $s^2 + (1+xA)s/Q + 1$, so the effect of increasing x is to reduce the Q to a new value Q' , equal to $Q/(1+xA)$, which explains the curves in Fig. 5(a). An analogous argument can of course be developed to explain the symmetrical form of the corresponding cut curves when x is close to 1.

Whether or not the behaviour in Fig. 5(a) is desirable is a debatable point, but fortunately one can have it both ways! Suppose the feedback end of R_{19} is grounded instead of being connected to the output of IC_2 . The circuit will now only boost, and the transfer function will be $-(D + (1-x)AN)/D$. Since A is large, the transfer function can be approximated by $-(1-x)AN/D$ except when x is close to 1, so the major effect of changing x is now to change the centre-frequency gain without affecting the Q . The response curves obtained under these conditions are shown in Fig. 5(b).

There are several ways of modifying the circuit to obtain these curves, and the method used is to some extent a matter of personal choice, but here are three! First, changeover switches could be used to ground one or other end of R_{19} to obtain either the boost or cut curves. Second, the gain of IC_3 could be made variable, when the curves in Fig. 5(b) would be obtained with R_{19} at maximum boost. The third possibility is my personal favourite, and I have indicated it on the circuit diagram (Fig. 3). This is to use a centre-tapped control for R_{19} (I really must apologise for continually recommending obscure potentiometers!) and to ground the tap via S_2 to obtain the Fig. 1(b) curves. The advantage of this method is that the boost/cut setting is determined only by the control setting, just as before, although the control law will be changed. As will be appreciated from the change in the form of the transfer function, the centre-frequency gain will approach 0dB less rapidly as the control

State-variable filters

Although the present circuit uses the state-variable approach to provide only a bandpass filter, high-pass (HP) bandpass (BP) and low-pass (LP) outputs are available simultaneously, as indicated in the accompanying derivation. Note that the basic form of the transfer function is quite simple, and the final expression is relatively cumbersome only because of the form of the a_1 and a_2 coefficients. The derivation also shows more clearly how it is possible to change the Q independently of the centre-frequency gain. Since $a_2 = 1/Q$, we merely have to vary a_1 and a_2 together, which is achieved in the present circuit by a variable resistor (R_{20}) to ground from the a_1 and a_2 summing point. This obviously requires that the two signals go to the same amplifier input, and since the a_2 coefficient must be positive, a_1 has to be as well. If this facility is not required, a_1 could of course be either positive or negative.

A further advantage of the state-variable approach is that it can provide any second-order function, although this has not been exploited in the present circuit. The HP, LP and BP outputs are summed by a further amplifier (see ref. 4 for the system transfer function), which allows the corresponding reject functions to be synthesised. By making the appropriate coefficients variable, it would be possible to generate a continuous range of bandpass and band reject functions within the filter itself, rather than by changing the position of the filter within an amplifier feedback loop as in the present circuit. There may not be much to choose between the two methods, but I preferred the feedback loop method because it can be used with

any kind of filter, and any number of filters can be placed within a single feedback loop as shown in Fig. 6. It also allows the choice of two sets of frequency response curves (see Fig. 5), which may not be so easy to arrange by the other method.

$$BP = HP \times -1/R_5 C_1 S, \text{ where } S = j\omega$$

$$LP = BP \times -1/R_6 C_2 S = HP \times 1/R_5 C_1 R_6 C_2 S^2$$

$$HP = a_1 \times \text{input} + a_2 \times BP - a_3 \times LP$$

$$a_1 \times \text{input} = HP - a_2 \times BP + a_3 \times LP =$$

$$HP \left(1 + \frac{a_2}{R_5 C_1 S} + \frac{a_3}{R_5 C_1 R_6 C_2 S^2} \right)$$

$$HP \text{ input} = \frac{a_1}{1 + \frac{a_2}{R_5 C_1 S} + \frac{a_3}{R_5 C_1 R_6 C_2 S^2}}$$

$$= \frac{a_1 R_5 C_1 R_6 C_2 S^2}{R_5 C_1 R_6 C_2 S^2 + a_2 R_6 C_2 S + a_3}$$

Referring to Fig. 2, the a coefficients are

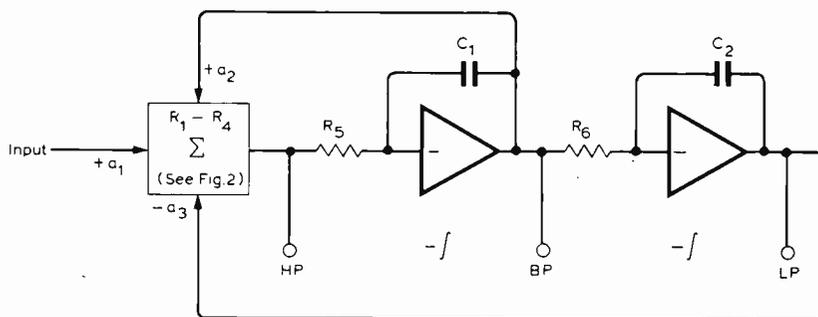
$$a_1 = \frac{R_2(R_3 + R_4)}{(R_1 + R_2)R_4}; \quad a_2 = \frac{R_1(R_3 + R_4)}{(R_1 + R_2)R_4}; \quad a_3 = \frac{R_3}{R_4}$$

Thus the complete highpass transfer function is

$$\frac{HP}{\text{input}} = \frac{R_2(R_3 + R_4)}{(R_1 + R_2)R_4} \times \frac{R_5 C_1 R_6 C_2 S^2}{R_5 C_1 R_6 C_2 S^2 + \frac{R_1(R_3 + R_4)}{(R_1 + R_2)R_4} R_6 C_2 S + \frac{R_3}{R_4}}$$

The bandpass and lowpass transfer functions are obtained by multiplying the high-pass transfer function by $-1/R_5 C_1 S$ and $1/R_5 C_1 R_6 C_2 S^2$.

When $R_5 C_1 = R_6 C_2$, $a_2 = 1/Q$.



is rotated towards its midpoint when the centre tap is grounded. Well, you can't have everything!

This effect can be reduced, however, by connecting a 1kΩ resistor between the slider of R_{19} and ground when the centre tap is grounded, which will mean using a double-pole switch for S_2 . The compensation is not exact, but it reduces the worst-case centre-

frequency mismatch to below 3dB. The ultimate solution would be to replace R_{19} by two parallel chains of resistors, one of which is grounded at the centre, and to select a point along one or other chain by a multiway switch. The resistor values would be chosen to obtain equal dB steps between the switch points, and it would probably be much quicker to determine the correct values

by measurement than by calculation!

We can now consider the remaining aspects of circuit performance. When LM318 devices are used, the distortion is extremely low, and it was difficult to make any reliable measurement at midfrequencies. For a +20dBm (22V peak-to-peak) output signal at 20kHz, however, I managed to obtain a value of 0.015%, but this fell rapidly as the signal level was reduced. In general, the control settings affected the distortion only insofar as they changed the output signal level. This excellent performance is a result of the very high bandwidth (15MHz) and slew rate (70V/μs) of the LM318, but there is sufficient latitude to allow the use of other devices for many applications.

The best alternative devices are the various families of f.e.t. input high bandwidth operational amplifiers, and the circuit performance was also evaluated with one of these, namely the Fairchild μAF356, which has a 5MHz bandwidth and 15V/μs slew rate. Using this device throughout, the distortion for a +20dBm output at 20kHz rose to 0.05%, and when the Q was increased at high centre frequencies, the centre-frequency gain also increased slightly – an effect not observed with the LM318. Device substitution showed that the effect, which occurred only at high Q, originated at IC₄, and most of the extra distortion was generated by IC₃. Both the LM318 and the μAF356 have an input voltage noise of around 15nV √Hz at midfrequencies, but the μAF356 may be slightly quieter since its input current noise is lower. I have not given any noise specification for the circuit, since the amplitude and frequency content of the noise will be greatly affected by the control settings, and to quote one or two blanket values could be misleading. However, I have tried to keep circuit impedances below 10kΩ wherever possible in order to keep the noise down to a level where it should be dominated by that of the ICs.

As mentioned previously, the circuit could be made more compact by the use of dual or quad i.c.s. A possible i.c. is the Texas TL074 series, but the bandwidth is only 3MHz, which will limit the performance at high frequencies. By the time this article appears, however, a wider range of quad devices may have become available.

Many applications will call for the use of more than one equaliser section, and the sections can be combined in two ways. The easier method is to connect them in series, and if the connection is permanent the buffer stage IC₁ can be omitted from the subsequent sections. This approach is best suited for a modular design, as it allows each section to be used independently. The other method is for the filter sections to be connected in parallel (as for a graphic equalizer), where IC₃-IC₆ and all the controls are duplicated, but share the same IC₁ and IC₂. The circuit configuration must be changed, however, to

allow the filter outputs to be combined, and the modified circuit is shown in Fig. 6. Circuit IC₂ is now a virtual-earth mixer, which can sum any number of filter outputs without interaction, but to achieve this the outputs have to be sent to the inverting instead of the noninverting input of IC₂, so we need to make a compensating phase reversal in the filter path. In theory, this could be done by moving the filter input connection from the noninverting to the inverting input of IC₄, but we would then lose the interaction which allows the Q to be varied independently of the centre-frequency gain (see appendix). The solution adopted is to rewire IC₃ as an inverting amplifier, which has the minor disadvantage that the gain of this stage will interact slightly with the setting of R₁₉, but the effect will make no difference in practice. Each filter section can be switched out independently as shown in Fig. 6, or they can be switched out together by a single switch between the common ends of the R₅ resistors and the inverting input of IC₂.

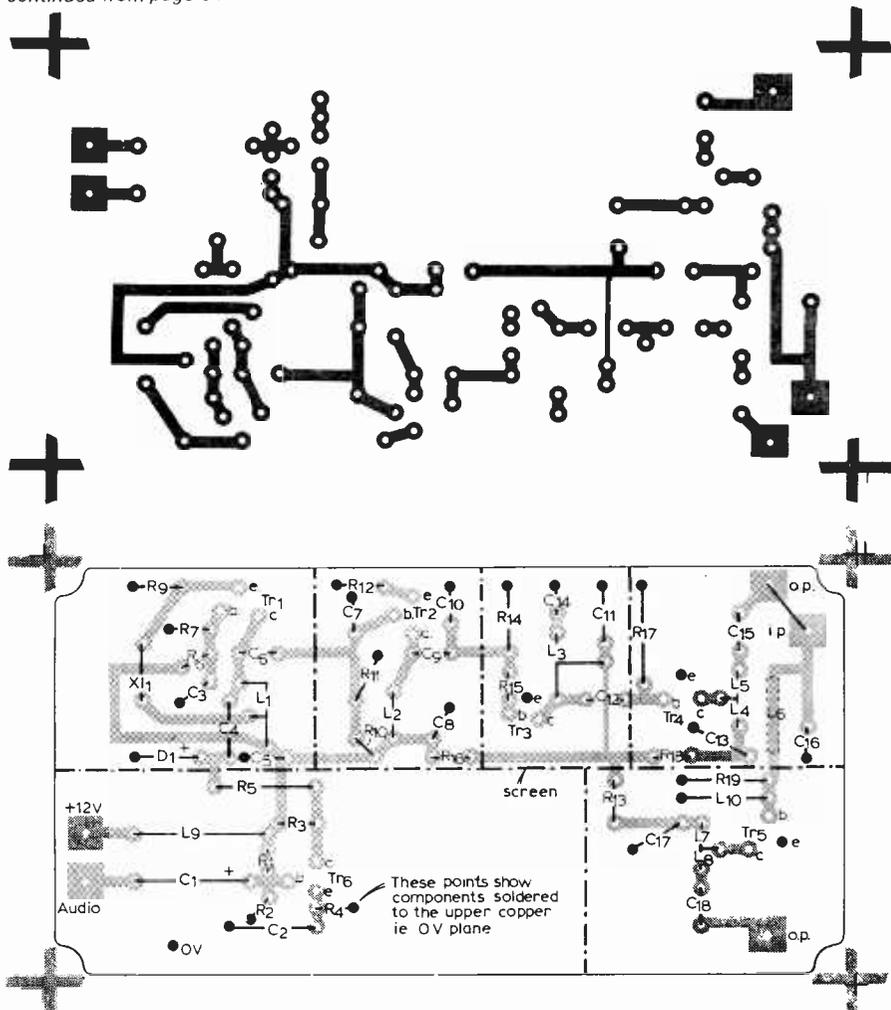
How does the circuit sound? My advice is to build it and find out! At low Q, the response can be corrected over a large frequency range by as few as two stagger-tuned sections, and in this mode the circuit is a very useful "shelf" filter. As the Q is increased, the circuit becomes more like a graphic equalizer, and ultimately resembles a musical in-

strument! A wide variety of special effects can be created by tuning one or two high-Q sections up and down the audio band, and at high Q the circuit also becomes a useful notch filter. Obviously this design is too complex for it to pose a significant threat to the popularity of the Baxandall tone control, even though it is a lot more versatile. But if you really prefer the mode of action of the Baxandall circuit, don't worry – this design will give quite a reasonable approximation to it if you tune one section to each end of the audio band and set them both to minimum Q. Now all you have to do is to label one control bass and the other one treble. Well, I told you it was versatile!

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50 years of "Empire" broadcasting

Each year the callsign of the late Gerald Marcuse, G2NM, is re-activated by the Chichester club to commemorate the many facets of his remarkable pioneering activities that extended over the period from about 1912 until his death in 1961. This year the event emphasised that it was Marcuse who during the period 1927 to 1929 provided the first series of broadcasts from the UK aimed at listeners in many parts of the "British Empire." These began in September 1927 some months before the first experimental BBC service from G5SW at Chelmsford and several years before the official start of the old BBC "Empire Service" in December 1932. *Wireless World* played a prominent part in campaigning for the broadcasts, against BBC opposition.

With Post Office permission, Marcuse broadcast daily from his home in Caterham, Surrey, including concerts and song recitals from a "studio" set up in the home of Percy Valentine and also unofficial relays of BBC medium-wave programmes. Even full-sized orchestras were fitted into the studio and many well-known musicians and singers took part. "Outside broadcasts" included bird songs from his garden and it was the G2NM broadcasts that first enabled listeners in many parts of the world to hear Big Ben. With a 100ft mast and Zepp aerial he ran about 1.5kW on 32.5m and the station was well received in many parts of the world until Post Office permission was withdrawn in 1929 and the role of Empire broadcasting was left to G5SW.

A British Oscar?

Almost a decade ago an attempt was made by a number of British amateurs to plan the construction of an amateur satellite for inclusion in the Oscar series ("Project Trident"). Although little came of these proposals, the idea has been revived, this time with the emphasis on providing the more technical amateurs with an experimental facility rather than a purely communications aid. The project is being formulated jointly by the University of Surrey UOS-AMSAT group and AMSAT-UK and will be run by the university's Space Studies Research Group, relying on support from industrial and research organisations and with the aim of establishing AMSAT-UK as a flight hardware group in its own right.

A number of suggestions on useful experimental facilities that could be included in such a satellite have been formulated, including the provision of real-time information for h.f. operators on the state of the ionosphere by using h.f. beacons as "topside sounders." It is also hoped that WARC 1979 will make provision for amateur satellites to carry beacons on microwave bands including



10 GHz. Martin Sweeting, G3YJ0 of the University group emphasises however that as a first venture the spacecraft would have to be kept simple and power consumption of all experiments might need to be restricted to an average of 5 or 10 watts.

WARC 1979

Although for most countries official proposals for frequency allocations to be formulated at the World Administrative Radio Conference next year still seem to be in a state of flux, radio amateurs have welcomed the news that the latest US proposals (although not necessarily representing final American policy) include three new amateur bands at 10, 18 and 25 MHz — 10.1 to 10.2 MHz, 18.068 to 18.162 MHz and 25.11 to 25.21 MHz — and in general represent an attitude favourable to the hobby. However, it is recognized that since the creation of the three separate ITU "regions" in 1947, amateurs in Region 2 (the Americas) have enjoyed significantly more favourable allocations than those in Region 1 (Europe and Africa) where European delegations have often proved among the most hostile to amateur allocations. The Home Office report "Preparation for the WARC 1979" (see July issue, News) notes that the decreased reliance on h.f. for international fixed service communications makes it possible to consider additional frequencies for various categories of users, including radio amateurs.

Spotty sun

Solar activity — and consequently maximum usable frequencies — continue to run ahead of predictions. This points either to an extremely high peak of activity during 1980, possibly even exceeding the remarkable Solar Cycle 19 peak of 1958 or to the peak being reached earlier than 1980. With several h.f. daylight radio blackouts this year, with transequatorial paths extending up to u.h.f. and the many auroral events

(averaging almost one day in three), it may well appear that we are already approaching peak conditions. Chris Bartram, G4DGU, however believes that the considerable number of 144MHz TE-mode contacts reflects the greater number of well-equipped 144/432MHz amateur stations resulting from Oscar satellite operations.

In brief

In an article in the UK FM Group (London) newsletter, Kris Partridge G8AAU proposed the introduction of 12.5kHz channel spacing in place of the current 25kHz in the f.m. simplex and f.m. repeater sections of the 144MHz band (145.0 to 145.837 MHz). This year's RSGB National Mobile Rally is at Woburn Abbey on August 6... Arthur Milne, G2MI, read his 1000th GB2RS news bulletin on May 7... The Home Office has resumed licensing of the "Phase 3" u.h.f. repeaters and will also consider applications for experimental repeaters on microwave bands although additional v.h.f. repeaters are still excluded... Special event "v.h.f./u.h.f." stations, with the prefix GB8, are being licensed by the Home Office through the RSGB... British amateur (maritime) licences for stations on board ship are no longer restricted to crystal control on h.f. bands... A beacon station, W6IRT, at Hollywood, California on 28.888MHz has been licensed for 6 months by the FCC... The International Amateur Radio Union is sponsoring a special amateur radio training course in Colombo, Sri Lanka with instructors from West Germany... Over 70 West German amateurs are operating on 10GHz. Activity is also reported from East Germany, Switzerland and Luxembourg... REF reports that 7 repeater stations (144 MHz) are in operation in France, 5 are undergoing tests; 5 are in construction; and 4 in the planning stage. Output ranges up to 100-watts and heights above sea level to 1,200 metres... According to Pierce Healy, VK2APQ in "Electronics Australia," an experimental amateur moonbounce installation of the University of Woolongong was wantonly damaged by vandals early this year. The station ("Project Dapto") has been built up over the past 8 years and may now have to be moved elsewhere... Evening classes for those taking the Radio Amateur's Examination in December or (with the new syllabus and with multi-choice questions) in May 1979 are being run in many parts of the country with enrolment during early September. Enquiries should be made at local adult education centres... Yukon, Canada is to use the prefix VY1... The Yeovil Mobile Rally is to be held at the STC/ITT Social Centre, Brixham Road, Paignton, Devon, on August 27.

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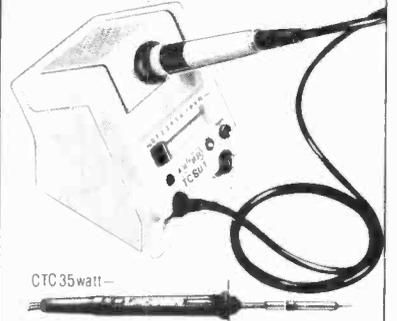
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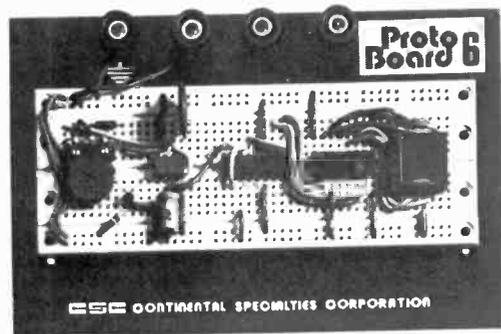
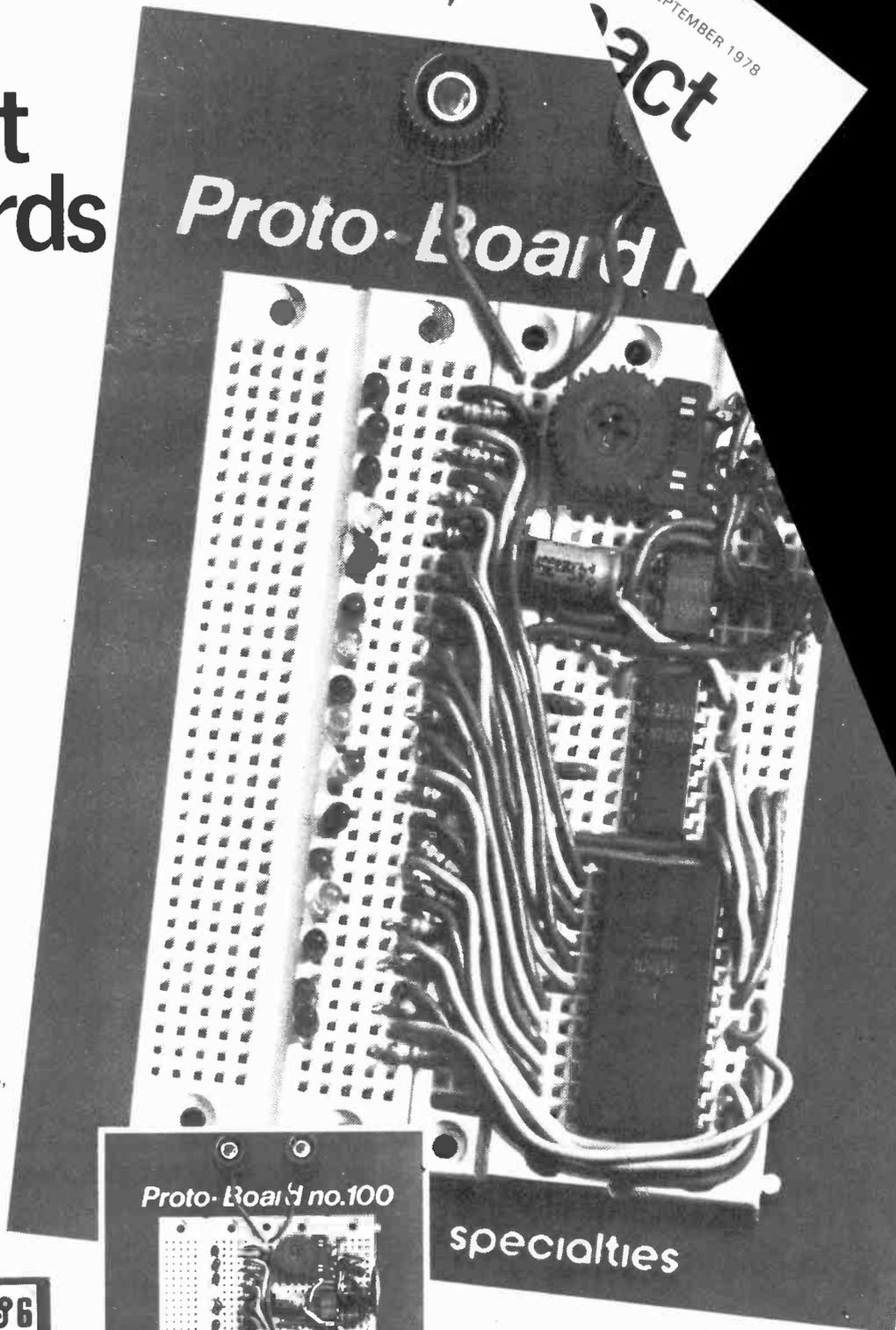
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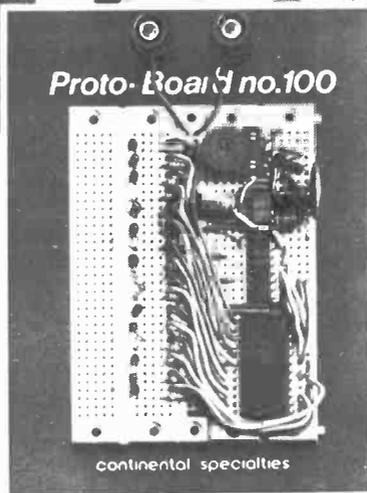
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Trends in microprocessors

An analysis of types now available on the market

by David A. Russell, B.Sc. Computer Technology Ltd

Since the last survey of microprocessors in *Wireless World* (December 1975 issue) a great many new devices with a wide range of capabilities have been introduced. This article provides a background to the current situation and discusses in general terms the directions that developments seem to be taking.

THERE are a number of starting points that could be considered when attempting to categorise the available devices, such as word width or technology. My own preference is to start from the product/market situation and determine where in the cost, performance and volume spectra the product will be. High volume, cost-sensitive applications will generally use a completely different type of microprocessor system from that used in a high performance, low volume application, even if both systems use 8-bit words. See Fig. 1, an adaptation of some information pro-

duced by Intel. Also, the memory and i/o requirements vary considerably, and this affects, for example, the type of memory used and the design of the i/o.

Before considering the details of the various configurations, it is worth looking at the costs of minimum sets of basic l.s.i. parts typically used in various microprocessor systems (see Fig. 2). The wide range of performances is reflected in a similarly wide spread in costs; it will be appreciated that, in practice, there are overlaps between the various cate-

gories shown. These costs do not include any allowance for overheads such as translators or buffers and drivers, printed circuit boards, power supplies and so on. Also, the quantities are assumed to relate to the application; for example, the minimum order quantity for single chip microprocessors is typically 1000-5000 pieces because they use mask-programmed r.o.ms., whereas 100+ volumes are shown for the high performance bit slice systems.

At the design stage, the integrated

	Chips	Quantity per annum
4-, 8- and 16-bit single chip microcomputers	1	5k - 1M
8-bit chip sets	2+	1 - 100k*
8-, 12- and 16-bit general-purpose systems	3	1 - 10k
16-bit high-performance systems	10	1 - 1k
2- and 4-bit slices	30	1 - 100

*Except automotive market

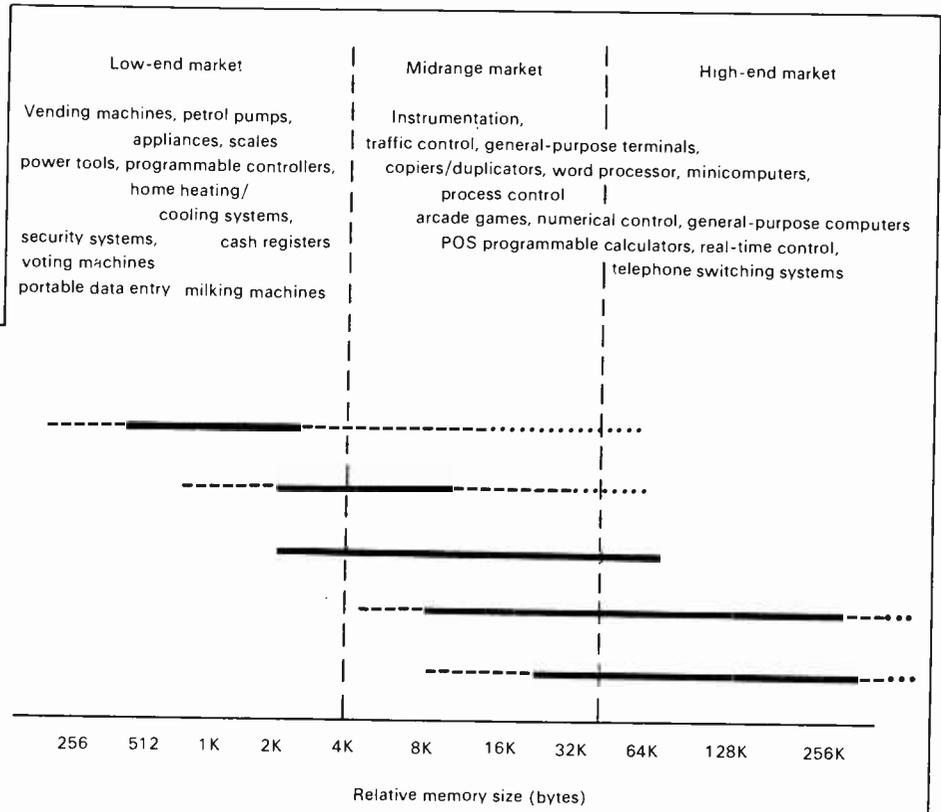


Fig. 1. Application markets related to types of microprocessors and relative memory capacity.

Explaining diagrams Fig. 4 onwards

In the diagrams, the rectangles drawn in thin lines are functional blocks, while the areas enclosed by thick lines represent actual chips. Where a functional block protrudes outside of a thick line, this means that extra logic, external to the chip, is required to take full advantage of the facilities available. Shaded areas imply that the part uses an interface specific to the microprocessor.

- Bit manipulation
- Decimal capability
- Timing control
- Tightly-coupled i/o
- Few peripherals
- Interrupt
- d.m.a.
- Relocatable programmes
- Intelligent i/o
- Many peripherals

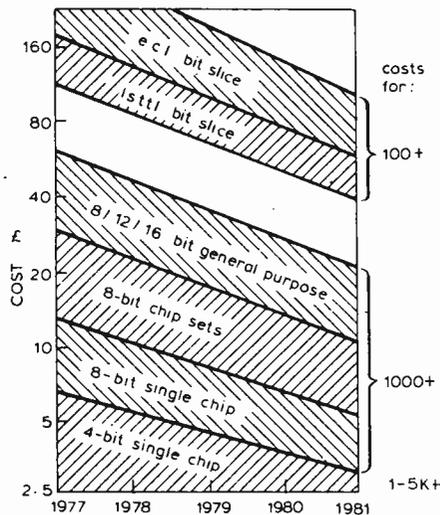


Fig. 2. Costs of l.s.i. parts used in microprocessor systems showing the falling trend over a number of years.

circuit manufacturers face specific trade-offs between costs and chip size, gate packing density, gate delays, power dissipation and package pin counts. To meet the requirements of the wide spectrum of applications the manufacturers are obliged to produce a range of products having differing implementations of the hardware functional blocks in a microprocessor system, depending on cost, performance, instruction set and flexibility (expandability) objectives. I shall illustrate this in the diagrams by using a standard format for the functional blocks (thin lines) and "overlying" the actual chip functions (thick lines).

The blocks generally include programme memory (r.o.m., e.p.r.o.m., or r.a.m.); data memory (r.a.m.); peripheral interface logic (general purpose or specific to a particular peripheral); timers (hardware timers are tending to replace software loops) and interrupt or test inputs. For high speed peripherals, a direct memory access facility is often included. These functions usually connect to the microprocessor unit (m.p.u.) via a common bus, and a clock generator and timing circuitry will control transfers across the bus.

Within the m.p.u. there will be an arithmetic logic unit (a.l.u.); the working registers (available to the programmer); internal registers (e.g. temporarily storing the current instruction or the next address); a control r.o.m. or equivalent logic, and instruction decode and sequencing logic (see Fig. 3). The established 6800 family is an example of a system in which the m.p.u. and the other functional blocks are provided by individual packages and the system is expanded by connecting more memory or i/o chips onto the bus.

Single chip microprocessors

The i.c. manufacturers seem to be agreed that a single chip microprocessor is a device that contains all the essential functional blocks (r.o.m., r.a.m., m.p.u., i/o, timer) to allow it to be

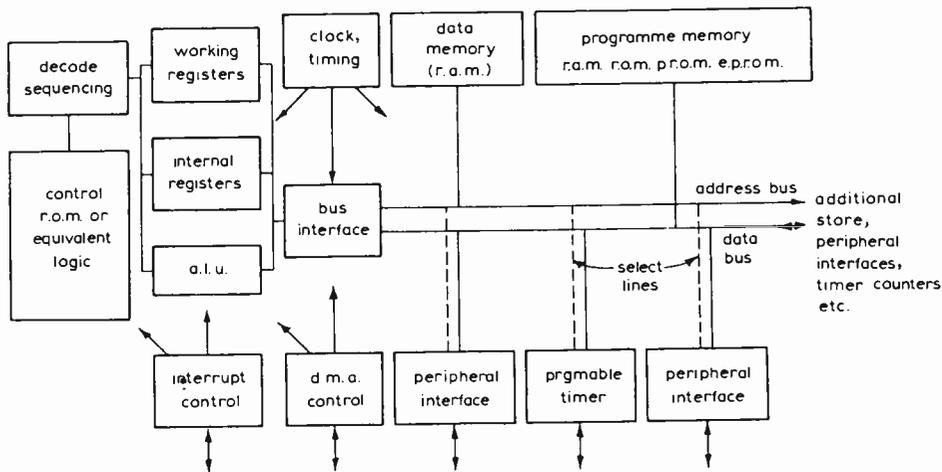


Fig. 3. Typical arrangement of functional blocks in a microprocessor unit.

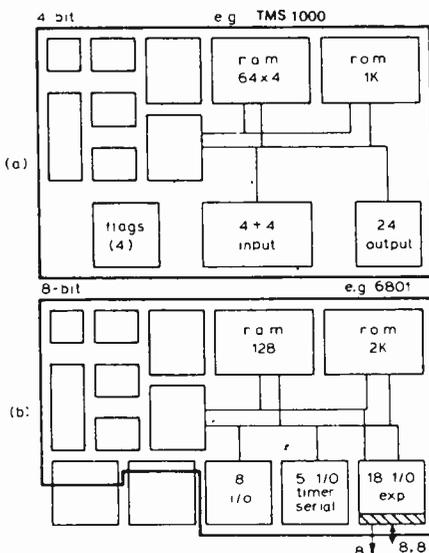


Fig. 4. Examples of single-chip microprocessors: (a) the 4-bit TMS1000, and (b) the expandable 8-bit 6801.

used in low cost, high volume applications such as microwave ovens, washing machines and electronic games. One can identify two basic types available: first, very low cost, non-expandable microprocessor chips, with fixed capacity of memory and i/o. These are devices like the 4-bit TMS1000 (Texas), the 8-bit 3870 (Fairchild, Mos-tek) and the more recent 8-bit 8021 (Intel), see Fig. 4(a). Because the applications are very cost-sensitive, the manufacturers are producing variations on the theme to meet particular requirements, with extra r.o.m. or i/o on larger chips, and, in the case of Intel's 8022¹, they have integrated much of the external logic normally required in microwave oven applications by including a two-channel analogue multiplexer and an analogue to digital converter on the chip. Intel say that this is the first of a number of 802X parts that will be designed for specific high volume applications.

The second type of single chip microprocessor is expandable, allowing the use of more memory and/or i/o than is

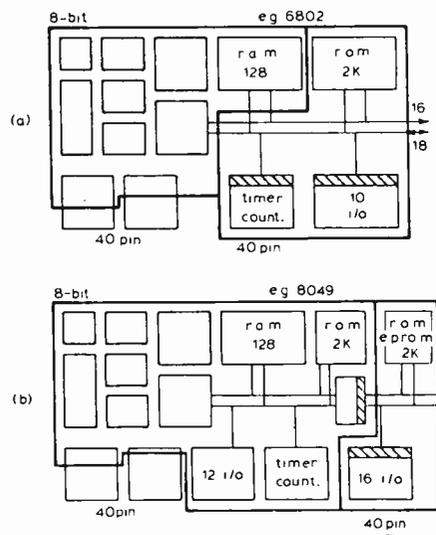


Fig. 5. Examples of two-chip set microprocessors: (a) the 8-bit 6802 and (b) the expandable 8-bit 8049.

included on the basic chip. This would also be useful where the design requires both r.o.m. and e.p.r.o.m.; the e.p.r.o.m. would allow specific customer variants to be produced, while the main programme would be in r.o.m. to reduce cost. Examples are the 8048 and 8049 families (Intel), 6801 (Motorola)², Z8 (Zilog) and 9940 (Texas). See Fig. 4(b) and 5(b). Some of these types are available with serial i/o for distributed processing, and in due course versions with e.p.r.o.m. rather than r.o.m. should be increasingly available, allowing low quantity applications to use single chip microprocessors. Minor variants of these microcomputers can be used as peripheral controllers on microprocessor systems such as the 8080, 6800 and Z80. Examples are the 8041 universal interface (u.p.i., Intel) and the 6801E (Motorola).

It is interesting to consider that the performance of the faster types of microprocessors exceeds that of the early 8-bit microprocessors, such as the 8080 and 6800, even though the faster devices contain so much extra logic!

Two-chip expandable systems

Another approach for obtaining flexibility is to base the system design on chips that split the minimum system into two packages and can be expanded by the addition of bus-compatible devices. The longest established example is probably the F8 (Fairchild). Others to consider are the 6802 (m.p.u. and r.a.m.) used with the 6846 (r.o.m., i/o and timer) from Motorola, the 6500 series m.p.us used with r.o.m., r.a.m., i/o and timer chips from M.O.S. Technology, and similar systems for the National Semiconductor SC/MP and Signetics 2650 (see Fig. 5(a)). More recently "cut-down" versions of single-chip microprocessors have been made available, such as the 8035, which can then be used with r.o.m., i/o combination chips or e.p.r.o.m., i/o chips from Intel.

As can be seen, with the introduction of the combination memory-and-peripheral chips, the various single chip microprocessors and the m.p.u., i/o and memory combinations, the designer has plenty of scope if r.o.m. based systems are required. The choice is more limited if e.p.r.o.m. is needed, but this situation should improve during early 1979.

8, 12 and 16 bit general purpose microprocessor systems

There are various situations where the previously mentioned systems would not be appropriate. For example, if greater performance is required or if large amounts of r.a.m. are to be used, such as in intelligent terminals or development systems, the familiar microprocessors such as the 8085, 6800, Z80 or 9980 would probably be the next types to consider (see Fig. 6). The families generally included selected high speed versions, with the manufacturers leap-frogging each other as new devices are introduced. The 8085A-2 and Z80A seem to be the fastest available at the moment (it depends on who is running

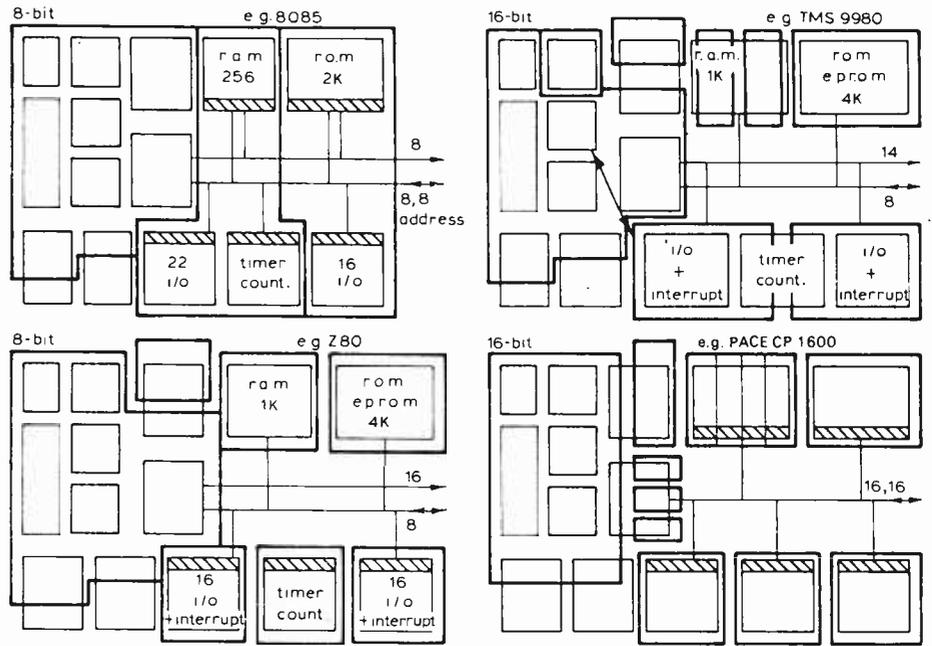


Fig. 6. Examples of general-purpose microprocessors: (a) the 8-bit 8085, (b) the 8-bit Z80, (c) the 16-bit TMS9980 and (d) the 16-bit PACE or CP1600.

the benchmarks as to which wins!). They will shortly be challenged by the 6809² (Motorola) which, like the Z80, has a large instruction set and extended register set, some features of which are described below (see Fig. 7).

The applications where these more powerful devices are used will often involve interfacing to a variety of peripherals, and the recent and continuing developments in peripheral controller chips are significantly reducing the design complexity, costs and chip counts incurred. Devices like s.d.l.c./h.d.l.c.* chips, floppy disc controllers, and c.r.t. controllers can replace a whole board of t.t.l. m.s.i. logic. Some of the more recent peripheral controller chips are actually based on universal peripheral interfaces (u.p.i.), so that the specific requirements of a high volume user can be taken into account by modifying the programme in the u.p.i. (e.g. the Intel 8278 matrix printer controller).

Another point to consider is that in some cases it is possible to use one manufacturer's peripheral chips with another's microprocessor, which may

be useful where your own manufacturer's device doesn't have the facilities required, or is not available.

The new 16-bit microprocessors

There is a lot of activity in the 16-bit microprocessors, with the 8086³ (Intel) and Z8000 (Zilog) coming onto the scene and the MACS (Motorola Advanced Computing System) in the design phase, to join the existing devices such as the 9900 (Texas), F100C (Feranti) and more recent 9440 (Fairchild). The 9440, 8086 and Z8000 are all claimed to have performances comparable with powerful minicomputers (i.e. Nova range, PDP 11/45 and PDP 11/70 without cache memory), and have very much larger instruction sets than most 8-bit microprocessors. In common with the 6809, the software features being emphasized by some manufacturers include the ability to use position independent code (which facilitates the use of r.o.m. libraries such as maths packages, interpreters and so on), the availability of a large number of registers, and instruction sets designed for array and repetitive operations, such as are required in compilers, editors and executives.

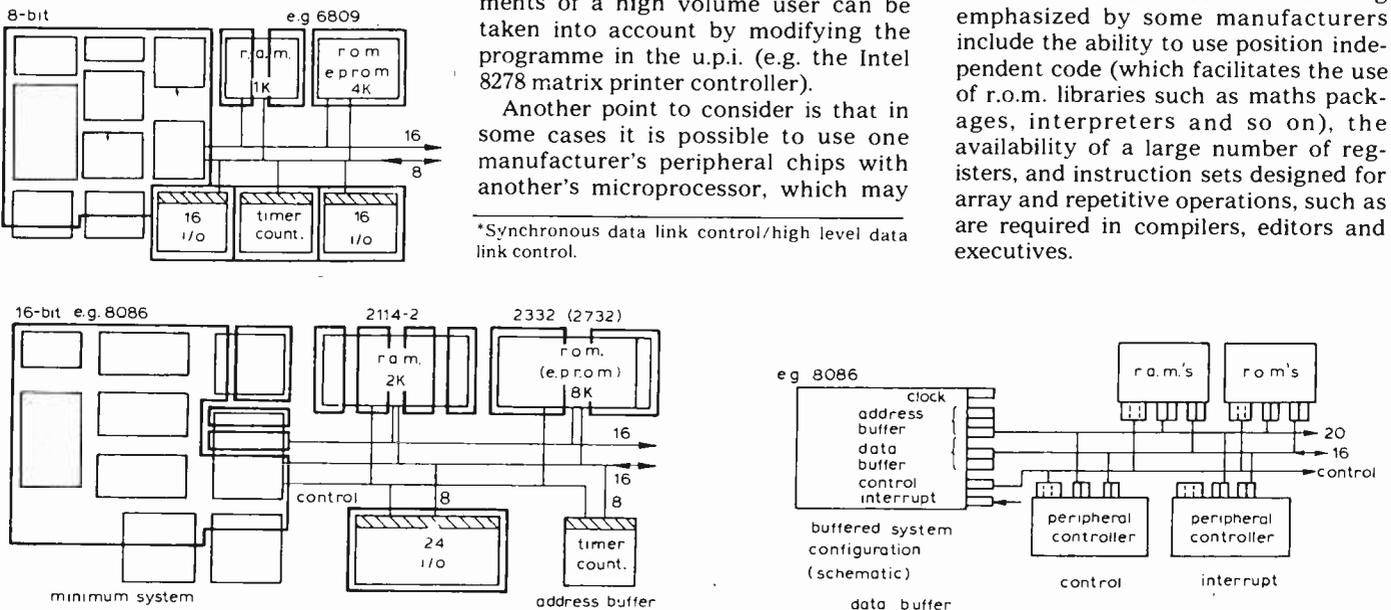


Fig. 7. High performance microprocessor systems: top, the 8-bit 6809; below, the 16-bit 8086 in the "minimum mode" and the "buffered mode".

*Synchronous data link control/high level data link control.

Another feature of some units is the inclusion of hardware and software controls for use in multi-microprocessor systems.

The 8086 and Z8000 are both able to address more than 64K bytes of memory, and to achieve the large address ranges both manufacturers use different configurations for small and large systems. The 8086 (1Mbyte addressing) has a pin that is strapped to V_{CC} or ground to determine whether the "minimum mode" or "buffered mode" is selected, while the 5Mbyte version of the Z8000 will use a 48-pin package instead of the 40-pin package used on the standard version (see Figs. 7b, 7c). Whilst such large address ranges may seem out of place in microprocessor applications, with the rapidly increasing capacity of memory chips and the advent of the r.o.m. libraries, addressing beyond 64Kbytes seems likely to become a useful feature in many applications.

Bit slice systems

The bit slice families have been developed as an extension to the existing Schottky t.t.l., e.c.l. and c.m.o.s. logic families, to combine the desirable performance characteristics and design flexibility of the logic families with the reduced costs and reduced package counts of l.s.i. Various chips are available that allow the designer to implement the functional blocks within a processor, such as a.l.u. and registers, control p.r.o.m., microinstructions sequencer (which determines the next address of the p.r.o.m.) instruction decode, and memory interface⁴.

In the schematic examples shown in Fig. 8 using the 2901 family (Advanced Micro Devices) the a.l.u., registers and some control is implemented in 4-bit slices, known as slice microprocessors. These can be cascaded to make a system of the desired word width; four of them are used to make a 16-bit system. The microinstruction sequencing is controlled by a set of chips that are also cascadable 4 bits at a time, although recently a single-chip sequencer has been introduced (2912).

The author

David Russell graduated from Southampton University in 1969 and completed a sandwich course at A.E.R.E. Harwell the following year. In 1970, he joined CTL, starting in the Circuits and Memories group. After working with semiconductor memories and high speed logic families he moved into the Systems group and was involved in the design of a number of products including power supplies, peripheral controllers and switching units for ultra-reliable systems. More recently in the Product Group, in which he is now the company authority on microprocessors, he has been working in applications using microprocessors in peripheral controllers, and has presented papers reviewing the microprocessor scene at several symposia in the last two years.

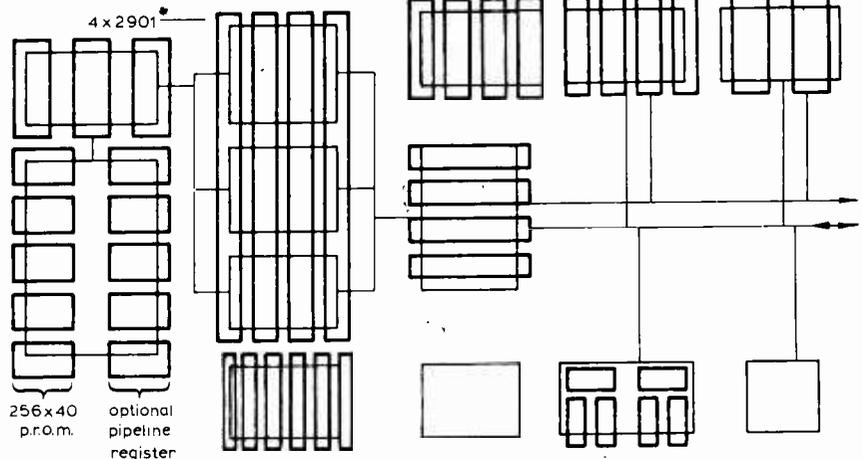


Fig. 8. Schematic showing the principle of bit-slice systems, here for example, 1 × 29811, 2 × 2911, etc.

These systems can, for example, be used for emulating existing minicomputers, or in the design of controllers for high speed peripherals such as rigid disc drivers⁵. The instruction set to be obeyed by the system is determined by the contents of the control p.r.o.m. The width of this is also in the hands of the designer, and may be in the region of 28-36 bits for small processors or peripheral controllers and 48-60 bits for emulation of powerful minicomputers.

The available families include the 9400 (Fairchild); 745481 (Texas) 6701 (Monolithic Memories) and the 10800 (an e.c.l. system from Motorola), but the market leader is the 2901 family, which has been very widely second sourced. A recent addition is the 2903, a 4-bit slice much like the 2901, except that more registers can be added onto the basic set via expansion ports and multiply and divide instructions are included.

To improve the performance of bit slice systems, more powerful memory control chips are being introduced that include an a.l.u. and registers, dedicated to calculation of the next address, while the main 4-bit slice microprocessor system continues with the current instruction. (This arrangement is also used in the 8086 16-bit microprocessor.)

As for future improvements in performance, the basic (internal) cycle times of Schottky t.t.l. systems probably cannot be reduced much below 150ns, so some of the i.c. designers are turning towards the use of e.c.l. circuitry inside the slice family chips, while retaining s.t.t.l. or l.s.t.t.l.-compatibility by putting buffers on the chips⁶.

An alternative that may become attractive to the minicomputer designer is to switch to using an e.c.l. bit-slice family to overcome the speed limitations of the other technologies, such as the 10800 (Motorola), especially as translators to s.t.t.l. bus systems are available, as are development systems.

Conclusion

When microprocessors were originally introduced they were cheap, but slow and very basic, requiring a considerable amount of support logic around them. We are on the verge of a

new phase, where the microprocessor manufacturers can provide practically tailor-made l.s.i. systems, for example at the high volume, low cost end, using the application oriented single-chip microprocessors, and, for larger systems, using peripheral controller chips and standard r.o.m. packages with high performance microprocessors. With the steadily falling cost and increasing performance trends, the point will soon be reached where conventional uses of microprocessors will leave a lot of power to spare, and new and novel uses for them will be devised.

A major problem the manufacturers now face is ensuring that designers are able to use the increasingly more powerful microprocessor in sufficient volume to justify the enormous cost of development.

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NEWS OF THE MONTH

Government conclusion —

The advantages of citizens-band radio are more than outweighed by the disadvantages

THE SUBJECT of citizen's-band radio was again raised in the House of Lords, on July 11, and again Lord Wells-Pestell made it clear that the Government had no intention of providing it in the UK. This time, however, his answers were far more relevant to the subject than they were last time (see News, July issue, p47).

Prior to the debate, Lord Torphichen told *Wireless World* that Lord Tanlaw would be posing a question in the House, and would be avoiding the use of the words "citizen's-band radio" — in the hope that it would be better received. Lord Tanlaw's question was as follows: "To ask Her Majesty's Government whether they will accept a recommendation of the National Electronics Council to improve public communications by allowing individuals access to the radio spectrum for A to B communication." Lord W-P, saying "No" and introducing the description "citizen's-band radio," replied that the Government remained of the view that the advantages of introducing such a service would be outweighed by the disadvantages.

To this Lord Tanlaw asked whether the Minister had conveniently overlooked that the radio spectrum ignored all national boundaries and was governed only by the law of nature, and if so could he justify his reply when there was no legal or con-

stitutional basis for any nation State to claim a part or the whole of the magnetosphere, or to prevent an individual from having access to it. Would he then say why the UK was one of the few democracies outside the Communist bloc that had not allocated a frequency over which members of the public could communicate freely with one another.

In reply, Lord W-P said that it was the view of the Government not to provide citizen's band radio for a whole variety of reasons—too many to go into at question time. The Government had taken advice and had looked at what had happened in other countries. "There are many competing demands by the necessary users of radio, by mobile radio and by commercial industrial firms."

Reminding the House that there was evidence of abuse and misuse in countries that have c.b. radio, he then made reference to part of an *Electronics Australia* editorial, printed in the RSGB's journal *Radio Communication*, which told of things that could be heard over the air, such as school kids swapping dirty yarns and prostitutes touting for business. The noble Lord did not think that this was funny and quoted another piece from the editorial: "It seems possible that citizen's band may even have played a key role in a recent murder." The final part of his quote said that c.b. radio was becoming

notorious, and many people were suggesting that the authorities should reverse last year's decision and try to suppress it altogether.

Lord Wells-Pestell concluded his answer by saying, "We see no reason to introduce the possibility of that kind of thing here."

When asked by Lord Tanlaw whether he was prepared to say that the examples which he had given did not take place over the telephone system, Lord W-P said that he did not think that the two were to be compared and pointed out that conversations on the telephone took place between two people and were not necessarily heard by a large number of people.

Lord Torphichen then wished to know whether the noble Lord, representing the Government, thought it wise that casual would-be users of radio communication should be forced to use either the already overloaded Post Office radio telephone network or, worse, to misuse the amateur frequencies. Seemingly becoming impatient, Lord W-P repeated that the Government had studied c.b. radio in other countries and had come to the conclusion that the advantages were more than outweighed by the disadvantages.

After the debate Lord Tanlaw told *Wireless World* that he will still continue to press for c.b. to be heard in the House of Commons. □

Mini-Nyquist speech prototypes in 18 months

CONTRACTS ARE already being negotiated for the commercial exploitation of a speech processing technique which can transmit at one-seventh of the Nyquist sampling rate. The technique, developed by Brigadier Reginald King with the collaboration of the School of Electronic Engineering under Professor William Gosling at Bath University, reduces speech to an "alphabet" of 27 "letters" which are then transmitted at about 1000 five-bit words per second. One of the problems yet to be overcome is that the samples appear at non-regular intervals and present transmission techniques allow only for regular transmission rates, but Brigadier King told *Wireless World* that a prototype device, using microprocessors, would be ready in about 18 months, and that commercial devices would appear in about four to five years if all went well. He emphasised that the technique did not make current techniques obsolete overnight.

Brigadier King, who completed his work during a sabbatical year at Bath University, said his work had been based on a well-known paper published many years ago in the United States in which the authors described the effects of severely limiting the amplitude of human speech. They discovered that with 100% limiting, when all that was left of the speech waveform was a series of events corresponding to the zero-crossing-points in the speech, the intelligibility was still 97%. This meant that although the odd

word was lost the sentences could still be understood. Brigadier King first worked on this and other speech processing techniques at the Royal Military College of Science at Shrivenham eight years ago.

The sound of such "infinitely-clipped" speech was, as Brigadier King says, "pretty awful to listen to," but it aroused a great deal of interest. King and others were sure that the time intervals between zero-crossing-points in human speech conveyed the bulk of the information, "but there was something missing; some mislaid clue that we had yet to discover."

Then in May last year Bell Laboratories published a paper explaining why researchers into zero-crossing frequency were barking up the wrong tree. The paper "proved" that the reduction of data rates by using zero-crossing was theoretically impossible. A lot of the other researchers turned to other things. King stuck with it. It was after this that he began a year's sabbatical at Bath. He stresses the value of Bath's co-operation. He was given the use of staff and a PDP8 computer. Just as valuable, though, was that he and Professor Gosling were agreed on their approach, and that Bath, too, had done some work on zero crossing.

"We were looking for a way of sampling without involving the Nyquist rate. We had got to dispense with amplitude descriptors and linear processing." The mathematical model was shortly provided in a book by two

Russians: "Distribution of Zeros of Entire Functions." Speech, said King, was an entire function, having real and complex components. "The thing that was missing was the locations of the complex zeros."

The task was to identify one sub-set of complex zeros which would identify speech. The technique was not entirely accurate, said Brigadier King, "but all modulation systems are approximations."

The work so far, using computer simulations, has only confirmed that the technique works. "We've only cracked open the oyster, as it were." Bath is now refining the method, identifying, perhaps, other sub-sets of complex zeros which might improve it. A key to the technique is that, effectively, the packets of five-bit words occur in regular clusters, and this makes further condensation possible. Although there are 27 letters in the current alphabet, alphabets with as few as seven or eight letters have proved intelligible, though unpleasant to listen to.

Brigadier King says the technique is much simpler than current vocoder techniques. According to an Army statement, the equipment could be sold "at less than a tenth of the cost of any other existing system and will be housed in a terminal smaller than a shoe-box."

The Army statement went on to say that the details of the technique were classified, but it appears that this is as much for commercial as for military reasons. □

Custom i.cs produced by computer aided design

EQUIPMENT MANUFACTURERS can now get integrated circuits custom designed and produced for them by a British computer aided design service and chip manufacturing plant which offers speed and convenience as its main features. Conventional draughtsmen's work in the layout of masks is eliminated and layout design time is reduced from weeks to minutes. Customers, it is claimed, can get a price quotation for a given number of completed devices on the same day that they bring in a diagram of a prototype system that is to be integrated. Finished samples of the manufactured devices are available in ten weeks. The convenience comes from the fact that a computer system can provide a quick feedback of information that enables the

customer to check the design process as it is taking place. For example, the computer will run a simulation of a customer's logic system to make sure that what is specified on the system diagram will actually do what the customer requires of it when it appears as a manufactured device.

Design it yourself

The new service has just been started by GEC Semiconductors Ltd, who specialize in custom designed i.cs, and is called 'Cellmos'. This name derives from the principle that the customer designs his own integrated circuit using standard "cells" or circuits from a library of circuits taken from standard 4000

series c.m.o.s. logic parts. At present the GECS Ltd library at their Wembley plant runs to about 500 items. Of course, this restriction of the customer's design options to a given library of elements is what "pays for" the gain in speed and convenience.

After a customer's engineer has designed his required system in these standard parts, and perhaps built a prototype in breadboard form, he supplies a logic diagram to GECS. At Wembley an accurate copy of this diagram is made and on it all the cells, or discrete logic circuits, and their connections are given code numbers and letters. From these code symbols a list is compiled which defines the logic completely in known terms and is a full description of the customer's requirements. In addition, the customer can supplement the circuit information with a set of test waveform definitions.

The list, either handwritten, typewritten or in computer readable form, is fed into a GEC 4070 computer to carry out compatibility checks (such as that an x-input gate is in fact receiving x-inputs). This may take anything from 15 to 30 seconds. If waveforms are supplied the service will run a logic simulation of the circuit on the same computer and return the results of this to the customer for him to verify that the circuit works as he intended and that the data have been transferred accurately. This takes from 1 to 13 minutes according to the complexity of the integrated circuit, and a similar time is needed to print out the results.

Once all this data has been verified by the customer's engineer and approved, it is released to the part of the service which lays out the integrated circuit. The layout process, claimed to be unique, optimises the placing of the cells and interconnections on the chip and then generates two plots on paper. One plot shows the proposed physical layout of the i.c. with pad positions and so on. The second plot is a diagram of the chip in logic diagram form, and allows the customer to check this version against his original circuit.

Checking the layouts

At this stage changes can still be made in the initial list to correct errors or modify the circuit. If this is done, the procedure is repeated until the customer's requirement is met. Once the two layout plots are certified correct by the customer's engineer, the layout is translated into magnetic tape format for the preparation of masks on computer-based equipment. If waveforms were supplied, these can be used to produce a test programme for an automatic test equipment used by the service.

The final cost of such a custom-made integrated circuit is determined by the chip area and size of package. The greater the number of cells, the greater will be the chip area, the larger the package and the higher the cost. But GECS claim that, because of the reduction in the time required for design and the convenience of the whole design approach, "the overall cost of developing custom l.s.i. circuits can be significantly reduced for the smaller quantity user."

NEB details confirmed

THE NATIONAL Enterprise Board has published further details of the newly-formed microelectronics company into which £50 million of public money is to be invested. A statement was issued on July 22 saying that, initially, £25 million was to be invested in a new company called Inmos, and an agreement to that effect had been signed by the NEB, Inmos, and the three founders: Dr Richard Petritz, Dr Paul Schroeder and Mr Iann Barron, of whom more later.

The NEB say that provision of the second £25 million will depend on the achievements of the company. The funding will be in the form of ordinary and convertible preference shares. "Key employees will have the opportunity to purchase ordinary shares in the company. When investment in the company reaches the level currently envisaged, the founders and the future employees could hold up to 27.5% of the voting shares in the company." The NEB's investment will have a preferred position because of the differentiation of rights attaching to the ordinary and preference shares.

Inmos will concentrate on the next generation of m.o.s. technology, according to the NEB. Their products will include very large scale integration (v.l.s.i.) memory and microcomputer devices. This means the production of 64K r.a.ms, compared with the current maximum of 16K. Inmos will also, it is hoped, produce microcomputers — central processing units (c.p.us) on a single chip.

The company's headquarters and production will be based in the UK but technological and product development will be split between here and the United States, where the biggest market is. Operations will start simultaneously here and in the USA. A prototype production line will be based in the USA but by 1981 Inmos plan to establish volume production in the UK. The first task will be to establish design teams and plan production facilities. An NEB spokesman said that, as yet, no other appointments had been made than the three founders, but that the company was having "discussions with people about recruitment." By the middle of the 1980s, 4,000 people would be employed in the UK and 1,000 in the US. The NEB said they were looking at sites though no definite decisions had been made. The statement said

that areas of high unemployment would be given special consideration, and the spokesman said this meant careful study of the North and North-West.

Dr Richard L. Petritz, 55, received a Ph.D. in Physics from North Western University. After lecturing and Navy work he was director of Texas Instruments' semiconductor r&d laboratory for ten years from 1958. He established TI's UK lab at Bedford. In 1968 he founded New Business Resources to launch new electronics companies and, in 1969, Mostek. As a consultant he advised the World Bank and the Korean Government in setting up electronic business in the Korean Republic.

Mr Iann Barron, 42, did Army and Air Force research after receiving a Cambridge MA. He was head of systems research in the computer research laboratory of Elliott Automation. In 1965 he founded Computer Technology Ltd, the first UK minicomputer company, and was managing director until 1971. Since then, as a consultant, he has advised the Department of Industry on future developments in computers and information technology.

Dr Paul Schroeder, 38, won a Ph.D. in Physics from Massachusetts Institute of Technology in 1967 and worked for Bell on memory design until 1974. He moved to Mostek becoming, in 1967, director of memory design engineering. He is thought of as a leading expert in m.o.s. dynamic storage devices design.

One of the most interesting aspects of these appointments is that the head of the trio, Petritz, has based his business in Dallas on the finding of funds for new ventures. It is a comment on the willingness of the holders of risk capital to take risks that he has now thrown in his lot with the NEB, though he is said to find the new venture attractive because it will eventually allow the company to regain its independence from the NEB. Another attraction is that labour costs in the UK are lower than in the US, a fact which has also encouraged Japanese investment here of late. The memory products, first off the production line, will be made in the US to build up the company, and the microprocessors will be made in the UK.

Post Office approve phone-line tv system — a new aid to the British police

THE POST OFFICE have accepted for evaluation an application by Aero & General Supplies, of Nottingham, for a slow-scan television (s.s.tv) system to be used as a private attachment to their public switched telephone network and private circuits. In addition, the British police, who were earlier given technical approval by the Post Office for a similar system for use on private circuits have now put s.s.tv into their research and development programme.

The heart of the system which has been proposed for the public switched telephone network is a slow-scan transceiver called the Robot Model 530. This unit is already in use in a number of phone-line tv systems in America and Canada, and has recently been technically approved in Spain and the Netherlands. According to Aero & General Supplies, the Post Office had four months previously similarly approved the system for use on private telephone networks in the UK, but policy issues delayed approval for the public network. One may be forgiven for speculating that these policy issues could have had something to do with the fact that a s.s.tv system of this kind could in many cases compete with the Post Office's proposed Viewphone.

The Model 530 s.s.tv system, however, must run the Post Office gauntlet—the usual process of assessments and trials — because the approvals are subject to it meeting their technical requirements, which are to ensure the system's compatibility with the PO's networks and systems. Robot are confident that these requirements, which should involve only minor modifications to the equipment, can be met. These modifications will be discussed later.

Both the police system and the proposed switched-telephone network system, which

are manufactured by Robot Research Incorporated of California, can be used with telephones, or any other "speech-communication" medium because they only require audio bandwidths to convey all of their picture information. The picture obtained is stationary and updated about every eight seconds (almost like a slide show)



Fig. 1

or can be held as long as required. The frame, in each case, is composed of a 256-line display having 128 x 128 discrete picture elements retained in a memory, and each coded into one of 16 grey shades. Although the picture definition does not compare with that of 625-line fast-scan tv, it is nevertheless surprisingly good, as shown in Fig. 1. A normal 625-line fast-scan tv camera is used to obtain the picture and this is sampled at a slow-scan rate and then transmitted immediately or recorded for later transmission if required. The display may be shown on a normal 625-line monitor, or even on a domestic tv receiver (slightly modified).

S.s.tv systems have created enormous interest throughout the world and are already being used by security firms, banks, police and meteorologists. They are, for example, ideal for the quick transmission of "mug shots," fingerprints, cheque signatures and for security surveillance. When connected to a telephone answering machine they enable one remote operator to contact any chosen premises, a bank for example, and see a picture of the strongroom within seconds. When a system is connected to a

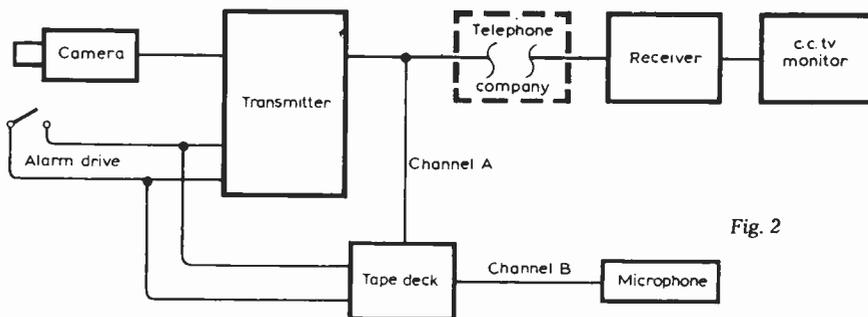


Fig. 2

More about s.s.tv

In s.s.tv a television picture is slowed down so that it may be contained within audio bandwidths. This slowing down results in a picture having about 120 lines (128 lines in the case of the Model 530) with a scan time of 7.2 (for a 120-line, 50Hz system).

Because the total bandwidth of s.s.tv lies well within the audio spectrum, it is possible to convey pictures using normal radio transmitters, telephones or other audio systems. In addition, the signals may be recorded on ordinary domestic tape recorders for later playback or for programme construction.

A slow-scan signal usually consists of a 1200Hz audio subcarrier which is frequency modulated by the composite video signal. The resultant f.m. signal is normally used by radio amateurs to modulate a s.s.b. transmitter. Figure 3 shows the frequency composition of part of the f.m. signal — a single slow-scan line — in which an audio frequency of 1500Hz represents a black level, and an audio frequency of 2300Hz represents a white level. Intermediate shades of grey are represented by the frequencies between 1500Hz and 2300Hz. In the case of the Model 530, each picture element is represented by one of 16 grey shades in a digital memory. The memory being a 65,536-bit store, made up of sixteen 4,096-bit r.a.ms.

The aspect ratio of a s.s.tv picture is usually 1:1, mainly because the surplus cathode-ray tubes, generally used by radio amateurs for s.s.tv, are round, and the square format used

the maximum available screen area.

All synchronisation pulses are transmitted at the 'ultra-black' subcarrier frequency of 1200Hz and consequently they do not appear on the screen. As shown in Fig. 3, the line scan consists of a 5ms sync pulse at 1200Hz followed by the frequency variations representing the light intensities of the visual image which has been scanned. The spot on the monitor screen flies back during the sync pulse period. At the beginning of each complete frame the 5ms sync pulse is replaced by a 30ms frame sync pulse during which the spot resets from the bottom right to the top left of the picture.

Although s.s.tv is almost entirely an amateur development pioneered by an American team headed by Copthorne Macdonald, W4 ZII, in 1958, it was only fairly recently (1968 in the USA and 1976 in the UK) that the controlling bodies (the FCC and the Home Office, Directorate of Radio Technology) put it into the standard radio amateur licence. Before about 1975, UK responsibility was with the Ministry of Posts and Telecommunications, who set fairly rigid standards for s.s.tv. These standards are no longer applicable, and providing s.s.tv is confined to the allocated frequency bands 3.5 to 3.8, 7 to 7.1, 14 to 14.35, 21 to 21.45, 28 to 29.7 and 144 to 146MHz, and the normal limitations of power and bandwidth are complied with, amateur s.s.tv transmissions may use any standards the operator wishes. This will not, however, be true of any unit proposed for private or public use with Post Office networks.

burglar alarm, the police can not only receive an alarm call, but can see a picture of what is happening. The eight-second interval between samples is sufficiently short for monitoring high security areas, and future developments are likely to include a system which compares one frame with the next to trigger an alarm after any picture-content change, caused by an intruder for example.

A typical s.s.tv system is shown in Fig. 2. This is a one-way system using transmitter and receiver separates. Transceivers can be used for two-way communications.

In the last few months, at least one British police force has carried out experiments with one of Robot's s.s.tv systems to determine its usefulness. They are using Robot's Model 400 which, being the amateur version of the Model 530, has more controls. As one might expect, they will be exploiting the system to the full and many modes and methods of transmission have been investigated — including telephone lines, and v.h.f., and h.f. radio — over both long and short distances.

For the system to be most beneficial to the police, they will be doubling the memory capability of the system and increasing the grey scale from 16 to 64 levels. Certainly, an s.s.tv system having 64 levels would be very useful for the transmission of both pictures and fingerprints. Robot also intend to make this modification some time in the future.

Continued on page 74

Because the Post Office insist on good mains isolation, one modification to the Model 530 will probably be to fit special transformers having screens between primary and secondary windings. The alternative is to add a mains isolation unit to each piece of equipment: this is how the police have solved the problem.

Problems are bound to arise because there is a Post Office line-signalling tone within the 1200 to 2300Hz bandwidth required for the s.s.tv transceiver. This tone, at 2280Hz, is used in the most common private (ac13 or ac15) and public (ac9 or ac11) networks to seize and release the line during phone calls. A 2280 ± 15 Hz receiver is used to sense this tone and it responds, only when the tone is pure and of sufficient duration, by cutting off the call. Since the s.s.tv system is f.m. it may be possible for it to produce a pure tone, which for one reason or another occurs within the bandwidth of the receiver and is, because of the picture content, of sufficient duration. This would also cut out the call.

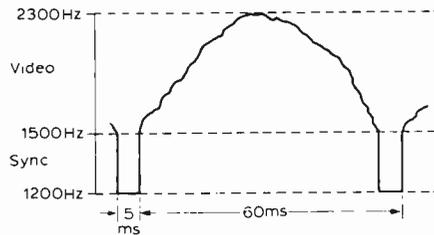


Fig. 3

Other tones on 2280Hz are for address signalling (dialling) and will therefore not affect the s.s.tv transmission.

One way in which this problem could be overcome is for the s.s.tv unit to produce a second tone (at 1000Hz say) so that even if a tone of 2280Hz was produced, it would not be pure because of the presence of the second tone — this would be interpreted as speech. Alternatively, the whole s.s.tv frequency band may be shifted down a few hertz to avoid 2280Hz. Robot considered the use of a

second tone but the idea was dropped when it was discovered that, whatever second frequency was chosen, the beat frequencies created either interfered with the picture or fell outside the frequency bands permitted by the Post Office. Eventually Robot chose a bandwidth from 800 to 1900Hz and are now trying to make this a standard acceptable to the whole of Europe and America. The police are also going to comply with this.

At the moment the model 530 is being retailed by Aero & General Supplies, but their intention is to distribute it through a franchised network consisting mainly of established closed circuit video dealers.

According to the company, a price cannot, at this stage, be fixed for the equipment if it is used on P.O. lines, but it is anticipated that a phone-line television transceiver, as it will be called, will be offered for less than £1,500. The equipment is available now for uses which do not require P.O. approval, for less than this, and transmitters and receivers for one way transmission are available for even less. □

Large-size l.c.ds to have longer lives

GLASS FRIT SEALING, currently recognised as the best method of sealing liquid crystal between glass plates for the production of long-life l.c.ds, has been so well mastered by ITT Components Group Europe that they are now successfully mass producing displays having character heights as large as 13mm. Few l.c.ds are given a life expectancy of greater than two years, but this UK company feels confident that their products will last for at least five years.

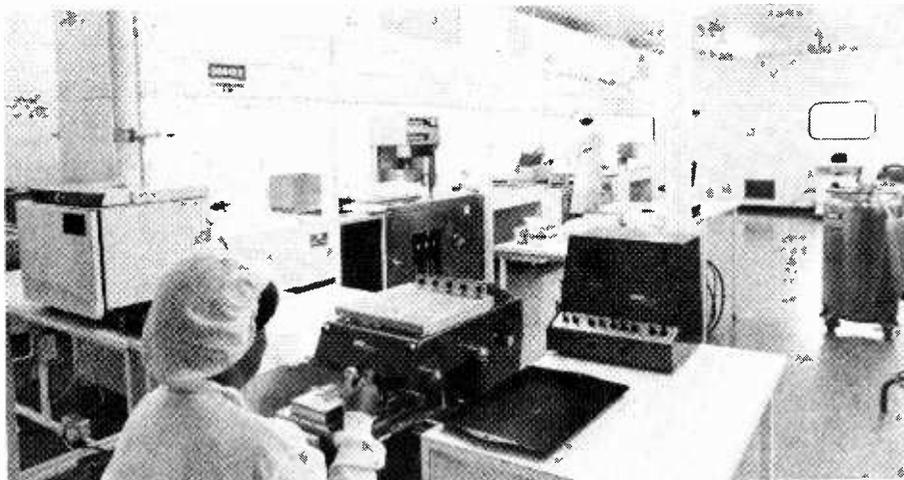
If a l.c.d. is to have a long life it is essential that the seal between the glass plates is impervious to all materials which could contaminate the liquid crystal. There are two types of seal in common use; glass frit and plastic. Since glass is more inert, physically and chemically, than plastic, it can withstand much worse environments, and is consequently expected to exhibit higher reliability. Unfortunately, the glass frit technique, which involves depositing low-softening point glass (frit), in paste form, on to the edges of the glass plates, and firing at about 500°C, is extremely difficult to master. Difficulties arise because the glass plates must be separated by only about 1/2 thousandth of an inch (12 microns), over the whole display surface and, of course, as the display gets larger these become even harder to overcome. Although a number of manufacturers are using the glass frit technique on small-size displays, Siemens in Germany, Brown Boveri in Switzerland, Electrovac in Austria

and Motorola in the USA, ITT claim to be the first company in Europe and probably in the world to produce large-size displays in quantity.

ITT, although late entering the l.c.d. market, recognised the long-term potential of glass-frit seals and decided, at an early stage, to concentrate on this technique. The real importance of the technique is expected to be seen in the future when the ambient temperature range of l.c.ds is extended. With the current, restricted, operating temperature range (typically 0 to 60°C) the difference in reliability between glass-frit seals and plastic seals is not marked.

ITT intends to achieve, within the next five years, a 25% share of the European market and a 10% share of the World market for displays having a minimum 12mm character height. The actual display that they are investing in is a field effective (twisted nematic) l.c.d. measuring 82 x 34mm overall. (For a description of the constructions and types of l.c.d. available see p230, May 1975 issue of *Wireless World*.)

General production area for ITT's liquid crystal displays. To achieve high yields it was accepted that airborne particles had to be virtually eliminated, and this purpose-built clean room area was constructed to create ultra-clean conditions for all critical operations.



The company is currently supplying l.c.ds at a rate of about 1,000 pieces per week, and they forecast that this figure will be up to 3,500 by the end of 1978 and 10,000 by the end of 1979. At the moment, however, ITT admit that yields are unrealistic and a figure of 60% is the most any manufacturer could expect to obtain. They are working to improve the performance of twisted nematic l.c.ds, in particular to extend their operating temperature range. In the future we can expect to see larger displays with drive circuits mounted directly on them.

At ITT's Central Research Laboratory at Harlow, completely new types of l.c.d. are being developed. These include a display based on cholesteric l.c. which contains a dye enabling it to have its own intrinsic colour and avoid the need for polarizers. Research is also progressing into the use of smectic l.c. materials in displays. These l.c.ds would have memory and continue to display a message after removal of the drive signal. □

News in Brief

The Radio Industries Club of Great Britain announce that Howard Thomas, Chairman of Thames TV, is to take over from Douglas Mugeridge, Deputy Managing Director of BBC Radio, as the Club's President. John Record, Thorn Industries' National Sales Manager, will take over from Alan Pederson, of Antiference, as the Club Chairman.

Admiral of the Fleet Sir Edward Ashmore, G.C.B., D.S.C., Chief of the Defence Staff until his retirement last year, has accepted an invitation to join the Board of Racal Electronics Ltd.

Robert Telford, Managing Director of GEC-Marconi Electronics Ltd, has been knighted for his services to export. Under Sir Robert's guidance the GEC-Marconi Electronics Group's overseas sales have risen from about £18 million in 1968 to over £230 million, and since 1966, the Group has been awarded 19 Queen's Awards, of which 11 were for export.

A Marshall (London) Ltd have moved their offices, sales and stores departments to new premises at Kingsgate House, Kingsgate Place, London NW6.

Mains interference and filtering

Protecting logic systems from mains borne noise

by I. Catt and M. F. Davidson (CAM Consultants), and D. S. Walton (Icthus Instruments Limited)

Although great trouble is taken when designing d.c. power supplies for large digital systems, interference on the mains power lines is often overlooked or underestimated. This article outlines the types of noise that occur, and describes a suitable filter for overcoming the problem.

INTERFERENCE from the mains can be classified into three types. Balanced, where the noise signal travels equally down the live and neutral lines, and the earth line acts as a return path. This is often called common mode noise, and it causes earth currents which can upset high gain linear circuits. Unbalanced, where the noise signal travels down the live line and back on the neutral line, leaving the earth line unaffected. This is often called differential mode noise and may be lost or suppressed in the d.c. power supplies of a circuit. It can be shown that any complex signal travelling down the three lines can be resolved into a common mode component and a differential mode component. Mains borne radiated noise, which can be both balanced and unbalanced, enters the equipment via the three power lines, and then radiates directly into the logic.

Susceptibility of a digital system to mains noise

Differential mode noise on the live and neutral lines tends to be smoothed out at the unregulated and regulated d.c. points. However, because large value capacitors have a significant series inductance, some of the noise, if not suppressed before the transformer primary, will pass through the power supply and cause transient variations which can disrupt the logic operation. Screening the transformer will not help significantly because differential noise is fed through the transformer from primary to secondary and not via inter-winding capacitance.

Common mode noise, however, does pass through the transformer via inter-winding capacitance, so a screened transformer will help to suppress the interference. The typical inter-winding capacitance for an unscreened transformer is 100pF. With screening, this falls to around 1pF. Any common mode noise that does pass through the transformer tends to raise the positive voltage relative to 0V, and tends to lift the

level of 0V at some points but not others. The use of a choke rather than a link between 0V and earth will help to render the logic immune to this noise because all of the logic supply lines will tend to move together. Therefore, any common mode noise which does pass through the regulated d.c. supply will see three loads in series. The link between the earth line and frame, the link between frame and 0V, and the line carrying 0V across the logic to the link. If the 0V to frame to earth link has a high impedance, such as a choke, most of the noise will appear harmlessly across it. The d.c. resistance of the choke should be below 0.1Ω to conform to BS3861. If, however, the 0V to earth link is a low impedance, the noise will lift the potential at one point on the 0V grid. This will degrade the logic signals and tend to cause a malfunction.

Mains borne radiated noise, which is emitted from the mains wiring, can be greatly reduced by screening the live and neutral lines, and earthing the screens to the frame. Another method of reducing the radiated noise is to include a mains filter at the point where the power lines enter the circuit module. A third approach is to have mains lines in the module separated from the vulnerable logic by correctly earthed bulkheads. Once past the mains filter, the mains cables do not normally need to be screened. If power is switched on and

off to loads within the module these power lines should also be screened.

Magnitude of mains borne interference

A reasonable noise amplitude to design against in a 240V single phase supply is 2kV over the range 100kHz to 10MHz. The noise may be common mode or differential mode and can be caused by, for example, switching off an electric motor which is on the same supply. It is wise to assume large amplitude noise above the nominal 240V of the line, and also that it is both common and differential mode.

The source impedance of the noise is difficult to determine, but it is safest to assume a very low source impedance of, say, two ohms. Both of these assumptions might surprise the reader, but they have been chosen to give a reasonable safety margin.

Mains filtering

Mains filters are constructed from capacitors and inductors. The capacitors require an adequate voltage rating and also have to be able to dissipate the heat generated from the maximum current. By Ohm's law, $V = IZ$ so $I = V/1/6 fC$ which is $240 \cdot 300 \cdot C$. Therefore, for a 1nF capacitor the current is around 100mA. It is worth noting that the mains filter can significantly alter the power factor of a load. The series inductance of such a capacitor can be as low as 10nH, which is very satisfactory in this application.

With inductors, it is important to make sure that they do not saturate at the peak current. If the power taken by a circuit is around 1kW, the r.m.s. current is around 4A and the peak current may be as high as 10A. A choke which saturates at 20A and has an inductance of 200μH can have a parallel capacitance as low as 10pF which again is satisfactory for this application. The d.c. resistance of such a choke is around 0.1 ohms, so it is possible to meet the safety requirements even if the choke is placed in the earth line.

A mains filter is a low pass device and the usual circuit is a double π as shown in Fig. 1. High frequency signals entering either the live or neutral lines see a high impedance inductor ahead and are shunted to earth through a low impedance capacitor. Typically, at

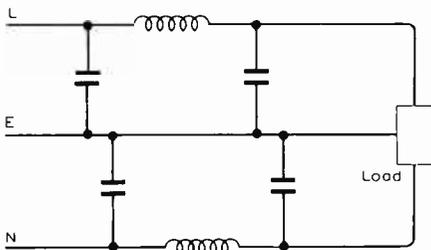


Fig. 1

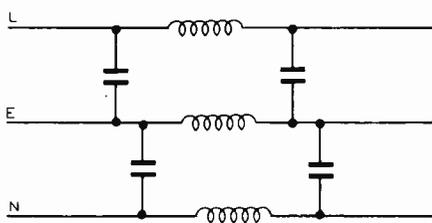


Fig. 2

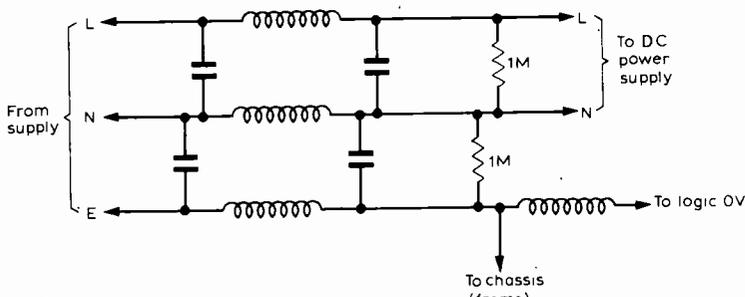


Fig. 3

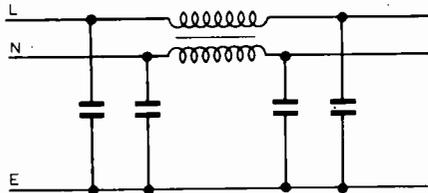


Fig. 4

1MHz, with $1\mu\text{F}$ capacitors and $200\mu\text{H}$ inductors, Z_c is around 0.2Ω and Z_L is around $1\text{k}\Omega$. If the source impedance of the noise is $1\text{k}\Omega$ or higher, the noise is attenuated by a factor of $1\text{k}\Omega/0.2\Omega = 5,000$ or about 70dB. If the source impedance of the noise is low, the first capacitor is ineffective, but the potential divider formed by the inductor and the second capacitor still gives around 70dB attenuation at 1MHz.

Any high frequency signals approaching from either direction see a short to earth, and a high impedance series inductor blocking the path ahead. This arrangement works well if noise is the only problem. But, because the input and output of both lines are connected together at high frequency, an "earth loop" pickup of externally radiated noise can occur. Also, the possibility of electrostatic discharge into the circuit is much more likely. From the point of view of radiated noise

the circuit in Fig. 2, which blocks the passage of high frequency signals down all of the lines, is preferable. It makes the path down the lines an open circuit to high frequencies, and tends to isolate the system. This filter does however cause a disquieting amount of earth current. If the capacitors are $1\mu\text{F}$, the total earth current is about 150mA. With the circuit rearranged as in Fig. 3, the noise suppression is virtually unaltered and the earth current is reduced to around 2mA. This circuit is also safer because there are no components linking live directly to earth, and a single shorted capacitor does not present a safety hazard. The two resistors discharge the capacitors if the filter is disconnected from the mains.

Commercial mains filters

Medium performance commercial filters have a specification of around

60dB insertion loss in the region of 1MHz. A filter of this type would cause 2kV of noise to be reduced to a mere 2V, which would easily be suppressed on its way through the power supply. Higher performance filters, specified at 100dB insertion loss, reduce noise of 2kV down to an unnecessarily low 20mV. The most serious shortcoming in commercial units is when the windings of both chokes are on the same core as shown in Fig. 4. The theory is that the currents in the live and neutral lines, being equal and opposite, create zero total magnetic flux in the choke. This means that for a heavy live and neutral current the core will not saturate, and a single toroid can be used in place of two separate and more expensive chokes. However, instead of two chokes there is a transformer which will not stop any differential mode noise. The author has not seen a manufacturer's specification where insertion losses for both differential and common mode noise have been unambiguously defined. The insertion loss is the ratio of the output amplitude from the filter into a load divided by the input from a source with the same impedance. Sometimes the specification does not state this impedance so it is virtually impossible to determine the filter performance. The correct specification is the minimum insertion loss when the source impedance and load impedance are independently varied from zero ohms to open circuit. □

This article is based on material from a book "Digital Electronic Design", by the above authors, published by C.A.M. Publishing, 17 King Harry Lane, St Albans, Herts, price £8.00 including postage. For information on the availability of the mains filters described write to Icthus Instruments Ltd, Princesway, Team Valley Estate, Gateshead. C.A.M. Consultants will be giving their next seminar on digital electronics design in St Albans, October 9 and 10. Information from 17 King Harry Lane, St Albans, Herts.

New 3-D military radar

MODERN RADAR DEFENCE SYSTEMS require faster reaction capabilities in order to combat the dramatic increase in aircraft speeds and manoeuvrability of targets. To achieve this, automatic data processing facilities, capable of carrying out high speed tracking and prediction from radar returns, have become essential. To operate with maximum effectiveness these automatic systems require continuous height information on all targets as well as their plan positions. Marconi Radar Systems Ltd have introduced a 3-D radar, called Martello, which does just this.

Martello has been designed for long range cover, transportability and to provide frequent height measurement to ensure that tactical height changes can be detected in good time. It is also equipped with full electronic counter-countermeasures (e.c.c.m.).

The radar, which operates in L band and provides automatic detection and plotting of targets, even in hostile environments, detects intruders at ranges in excess of 300 nautical miles and altitudes in excess of 100,000 feet. Elevation cover extends from zero to 30 degrees.

Range, azimuth and height is recorded for every target detected, on every revolution, with accuracies of 0.05nm, 0.5nm (at 100nm) and 1,000 feet (at 100nm) respectively. For height finding Martello uses what is claimed to be a unique parallel receiving system. This has a vertical stack planar array antenna comprising sixty identical horizontal linear array elements, each with its own receiver. Each array has the same-shaped amplitude distribution, giving a narrow azimuth beam width. By precisely controlling the amplitude and phase feeds to each array, the side lobes

are kept to a minimum. In elevation, the transmitted r.f. power is distributed between the arrays to give cos^2 target illumination, and every target within the elevation coverage is illuminated on every transmission.

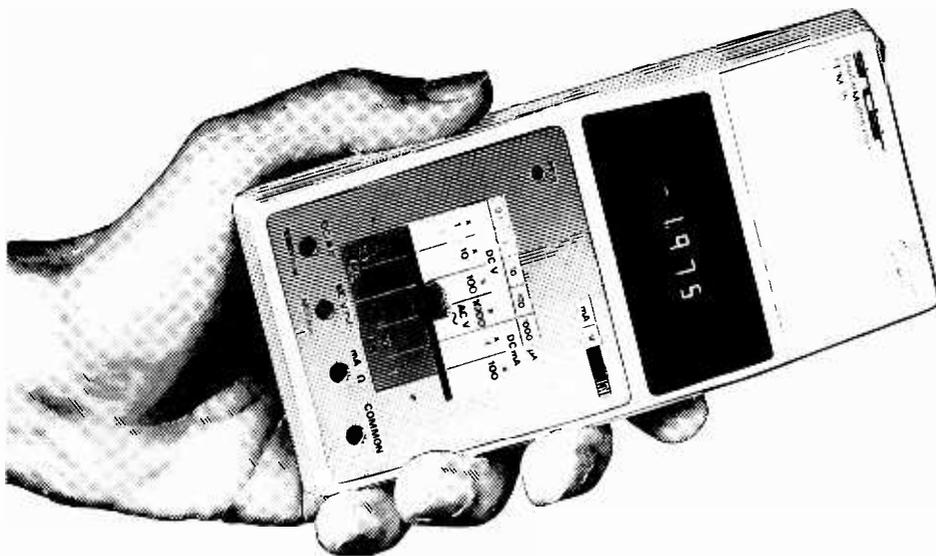
Returns from a target are received by all of the arrays and the individual receiver outputs are then combined in a simple passive beam-forming network. This synthesizes the cos^2 surveillance pattern and eight elevation patterns matched to the required elevation coverage. The surveillance pattern and the lowest elevation pattern are pulse-compressed and processed either automatically or manually. Target range and azimuth are extracted from series of individual returns by a plot-forming unit and the height data is obtained by measuring the returns in adjacent elevation patterns.

The system is designed to self-adapt to the radar environment and it has comprehensive facilities for monitoring system performance necessary for complete control of the system parameters.

The Sinclair PDM35.

A personal digital multimeter for only £29.95

(+8% VAT)



Now everyone can afford to own a digital multimeter

A digital multimeter used to mean an expensive, bulky piece of equipment.

The Sinclair PDM35 changes that. It's got all the functions and features you want in a digital multimeter, yet they're neatly packaged in a rugged but light pocket-size case, ready to go anywhere.

The Sinclair PDM35 gives you all the benefits of an ordinary digital multimeter – quick clear readings, high accuracy and resolution, high input impedance. Yet at £29.95 (+8% VAT), it costs less than you'd expect to pay for an analogue meter!

The Sinclair PDM35 is tailor-made for anyone who needs to make rapid measurements. Development engineers, field service engineers, lab technicians, computer specialists, radio and electronic hobbyists will find it ideal.

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What you get with a PDM35

- 3½ digit resolution.
- Sharp, bright, easily read LED display, reading to ±1.999.
- Automatic polarity selection.
- Resolution of 1 mV and 0.1 nA (0.0001 μA).
- Direct reading of semiconductor forward voltages at 5 different currents.
- Resistance measured up to 20 MΩ.
- 1% of reading accuracy.

Operation from replaceable battery or AC adaptor.
Industry standard 10 MΩ input impedance.

Compare it with an analogue meter!

The PDM35's 1% of reading compares with 3% of full scale for a comparable analogue meter. That makes it around 5 times more accurate on average.

The PDM35 will resolve 1 mV against around 10 mV for a comparable analogue meter – and resolution on current is over 1000 times greater.

The PDM35's DC input impedance of 10 MΩ is 50 times higher than a 20 kΩ/volt analogue meter on the 10 V range.

The PDM35 gives precise digital readings. So there's no need to interpret ambiguous scales, no parallax errors. There's no need to reverse leads for negative readings. There's no delicate meter movement to damage. And you can resolve current as low as 0.1 nA and measure transistor and diode junctions over 5 decades of current.

Technical specification

DC Volts (4 ranges)

Range: 1 mV to 1000 V.
Accuracy of reading 1.0% ± 1 count.
Note: 10 MΩ input impedance.

AC Volts (40 Hz-5 kHz)

Range: 1 V to 500 V.
Accuracy of reading: 1.0% ± 2 counts.

DC Current (6 ranges)

Range: 1 nA to 200 mA.
Accuracy of reading: 1.0% ± 1 count.
Note: Max. resolution 0.1 nA.

Resistance (5 ranges)

Range: 1Ω to 20 MΩ.
Accuracy of reading: 1.5% ± 1 count.
Also provides 5 junction-test ranges.

Dimensions: 6 in x 3 in x 1½ in.

Weight: 6½ oz.

Power supply: 9 V battery or Sinclair AC adaptor.

Sockets: Standard 4 mm for resilient plugs.

Options: AC adaptor for 240 V 50 Hz power. De-luxe padded carrying wallet. 30 kV probe.

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Model 178 offers functions and ranges for most measurement needs: 100 μ V to 1200V dc, 100 μ V to 1000V ac, 0.1 Ω to 20M Ω .

Model 179 is a full-function, multi-feature model offering the same advantages as the 178. Plus TRMS AC; 10 μ V Sensitivity; Hi and Lo Ohms, AC and DC Current. Yet it's still half the price you'd expect. Only £199.

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MOBILE RADIO BANDWIDTHS

It is with great interest that I read Mr W. M. Pannell's article on mobile radio bandwidths in the June issue. Some of the proof of the pudding has been tasted by us, a public utility, some five years or more. In the face of great scepticism we obtained permission to operate some standard USA made 30kHz fm fixed and mobile equipment at 15kHz spacing using a little more than 2kHz deviation. The equipment is used in hilly and mountainous country.

The practical result of our experiment, using single frequency simplex in the two-metre band, can be summarized quite simply:

1. Utilisation of speech under difficult conditions with no perceptible degradation of intelligibility.
2. If base stations (100 watt) are placed at least some 25km apart, possible use of the next channel at a 15kHz spacing.

Please note all this with standard 30kHz equipment. It is also necessary to state that propagation conditions in this area are such that they may almost be defined as anti-fading conditions, such that good results are a little surprising even to us. However, the peripheral mobiles suffer less noise (see par. 1) as a result and so operator acceptance was maintained.

Mr Pannell is to be congratulated on the redefinition of a technical problem which has very many economic and investment aspects. However, I hope that good sense will win out and that 12½kHz or 15kHz channels will become a standard. The latter seems the more likely to me, due to existing channel allocation in the USA and also its greater suitability for future 2.5kHz narrow band allocations (see June issue, p. 48; "FCC produces ideas for better spectrum use"), which may yet be interleaved with the existing allocation scheme.

P. Hirschmann

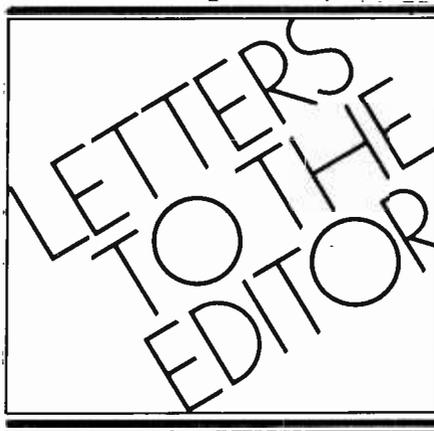
The Israel Electric Corporation Ltd
Haifa
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AUDIO EQUIPMENT REVIEWS

In a report in your July issue (p.49) you asked the question: "How accurate are audio reviews?" Instead of attempting to give the answer your reporter used his available space to explore the inner details of particular events. From this the reader could come to some extremely unpleasant conclusions. May I, therefore, through the courtesy of your pages and as a reviewer of long standing, attempt to supply the answer that you failed to give?

The accuracy of a review will depend upon a number of factors. Among these some of the most important are:

1. The personal experience of the reviewer himself which will obviously be variable from individual to individual.
2. The adequacy of his measurement facility to cope with the fine limits of distinction that must be observed in a scientific appraisal.
3. The ability of the reviewer to interpret correctly the integration of his measurements and subjective impressions, taking into account the further variations induced by programme source material, listening room acoustics and the fallibility of his own ears.



It follows that just as there is no "perfect" piece of equipment neither is there a "perfect" review. The eventual arbiter is in every case the customer, upon whom there devolves the final decision which is expressed through his cheque book.

As a matter of interest I have been using Bowers & Wilkins DM7 loudspeakers in my listening room for several weeks. My subjective impressions are slightly different to the opinions quoted from the *Gramophone* review by Mr John Gilbert, but more closely accord with his feelings than with the comments of Mr Attewell as expressed in your extracts from his *Hi-Fi News* review. This fundamental difference does not in any way invalidate either of the two published reviews but serves only to underline the dangers of confusing "accuracy" with "conformity".

It seems to me that your reporter was preoccupied with considerations of *integrity* rather than *accuracy*. Let me therefore add that I have in the past been pleased to accept hospitality from Bowers & Wilkins and in my capacity as publisher of *Hi-Fi Trade Journal* have indirectly accepted money from that company as payment for advertising space. The nub of the question is surely this: Have the "wicked men" of Worthing "got at me" to express the opinion quoted above? I find it very hard to believe that any serious minded person could entertain such a preposterous proposition for a single moment. The truth of the reviewing business is very, very different indeed.

In some twenty years I have frequently written critical reviews, some of which have been so condemning that the product has been withdrawn from the market. In all that time I can recall only one instance where any unpleasantness arose between the manufacturer/distributor and myself. To the contrary, an accurate but "bad" review is regarded as helpful by any sensible manufacturer since it points to the direction of essential improvement if he is to be successful in the market-place. This success is, after all, his final objective.

I have never at any time been put under any pressure to alter or suppress measurements; such a thing has never even been suggested to me and if it were the proposal would be most forcefully rejected.

In order to consolidate my own position as an "accurate" reviewer I have made a personal investment in test equipment and facilities amounting to a five-figure sum. My laboratory is used for consultancy work on behalf of clients, some of whose products I have reviewed or will be reviewing. The object of this exercise is to advise on performance characteristics so that improvement can be effected where necessary. Is this not in the true interests of both the consumer and

the distributor? To suggest any kind of malfeasance is to imply a contrary endeavour; i.e. a deliberate attempt to market a poor quality product to a gullible and unsuspecting public. Any firm which attempts to embark on such a course within this highly competitive industry would be taking the shortest possible road to disaster.

Perhaps my point of view might be regarded as naive, originating from a simple mind. If so, such a judgement will be accepted as a compliment.

Denys G. Killick
Pontypridd
Mid Glam.

WANTED: EARLY WW VOLUMES

The Radio Society of Great Britain is anxious to obtain copies of the early volumes of *Wireless World* for the period when it was the official organ of the Wireless Society of London and later the RSGB.

The volumes required are: volume 8, 3 April, 1920-19 March, 1921; volume 9, 2 April, 1921-10 March, 1922; volume 10, 1 April, 1922-30 September, 1922; volume 11, 7 October, 1922-31 March, 1923; volume 12, 7 April, 1923-26 September, 1923; volume 14, 2 April, 1924-24 September, 1924; and volume 15, 1 October, 1924-4 February, 1925.

The Society is also anxious to obtain volume 2 of *Experimental Wireless*, October, 1924-December, 1925.

If any of your readers could help the Society by supplying any of these volumes I would be glad to hear from them.

G. R. Jessop, G6JP
Consultant
Radio Society of Great Britain
35 Doughty Street
London WC1N 2AE

POOR PROSPECTS IN ELECTRONIC ENGINEERING

How wholeheartedly I concur with the perceptive exposition of your contributor in "Open letter to Finniston" (Letters, July). I, too, am saddened by our industry's attitude to young graduates. It is in the nature of the "small-minded people" who manage our industry that jealousy ranks high in their emotions, and leads to the notion that these arrogant young persons need to be kept in their place. If they are engineers that place will be fairly lowly. No lunches with customers or company cars for them!

The next hurdle for the unwary graduate I would describe as the Barnes Wallace syndrome, experienced as total opposition to the novel concept. The grinding down begins! This, surely more than any biological phenomena, leads to the misconception that a person's inventiveness declines asymptotically to zero at the age of 30. I satisfy myself that this is a myth by observing that my own "rate of acquisition of patents" index is no less now that I have passed 40. Not being afraid of the unconventional I have obstinately gone on designing things and being an engineer. But at what cost?

Taking the price of a house as reference level, my salary in a senior engineering post

is less now in real terms than it was over 20 years ago when I took my first job as a technical assistant.

To my mind, a crucial factor in this continuing decline in living standards is that engineers are not generally militant by nature. Even if they were, what hold have engineers got compared with, say, train drivers? If we all stopped designing and developing things tomorrow, who would notice? Who, anymore than would have "noticed" if radar, television or the digital computer had never been invented?

If it is accepted that the electronics industry has some merit in society, then for its survival very positive steps must be taken to provide a sufficiently attractive working climate for the young to develop their engineering skills, and without the necessity to go abroad, move into "sales" or "management" or even the drawing office, to feed their children.

*D. B. Brown
Charlwood
Surrey*

Several points raised in the "Open Letter to Finniston" by "Chartered Engineer" (Letters, July) have a depressingly familiar ring to them. Phrases such as "short-term industrial fodder" – "managed by narrow and small-minded people" – "irrespective of experience" – are amply justified in the light of my own, albeit short, experience.

To a young electronics engineer working at least, until recently – in a large company, part of an even larger privately owned group, the management philosophy (or is it a lack of one?) has been made abundantly clear. Staff who have gained experience on a particular system – and who are consequently much better at their job than those who have never seen that system before in their life, regardless of qualifications – are treated no better in any respect than new recruits. For instance, graduate entrants with no previous training are paid the same, in some cases greater, salary as some of my contemporaries who have not only had a year's system experience but also undergone "sandwich" course training with the company. And though they may start out as "bright eyed and bushy tailed", by the time they have grown wise to the way they themselves are being treated (even with the least perceptive engineer this process takes only a few months – in some cases less than a week), disillusionment and cynicism have set in for good – as witness the present writer. In such a climate, even if the job itself and its environment are pleasant and enjoyable – not always the case – dissatisfaction is inevitable.

If this malaise were confined to one large company, or even to one group, there would be little point in publishing this letter. But, from conversations with friends who have joined other organisations, this seems not to be the case. How many engineers in large companies do you know who are quite satisfied with their employers? The solution is therefore not to be found simply by moving among different leviathans (the grass is always greener...). A general rule would appear to be that the smaller a firm is, the better it treats its employees; but, of course, small firms do not employ the bulk of the country's electronic engineers, and also they prefer those who have already gained some practical experience. Hence the reluctance of graduates to enter the industry, an instance of which is provided by the letter cited above.

In these circumstances, the most sensible course for a young engineer is initially to take a job with a large company solely in order to gain enough experience that he can leave after a year or two, to join or form a small firm where he will have a much greater chance of finding job satisfaction. If this were to happen on a large scale, the effect would be to leave the large groups with inexperienced youngsters and experienced no-hopers (some would have it that this is already so); in which case, if their senior management were unaware or, through inertia, unable to rectify the situation, they would simply collapse, due to inability to compete – in the absence of other market factors. It may be that this process is already in motion, and that the elephants have had their day. While to many this would be a cause for rejoicing, it would be a shame if the vast resources inherent in them were to go to waste, and it is to be hoped that some at least of their creative potential could be salvaged in such event.

Young engineers should therefore be encouraged to be more aware of the state of the industry they are entering, and reassured that there is a future for them, albeit perhaps along slightly different lines to what is obvious at the present. It is, after all, far better to give birth to something new and vital than to despair of the dying.

*Tim Williams
Ely
Cambs.*

The "name and address supplied" open letter to Finniston in your July issue makes chilling reading. There, but for the grace of God, go I. The poor man is fifty-one, trapped in the horror of an electronic engineering career.

I graduated in Engineering in Cambridge in 1959 and went into electronics R&D. I finally read the writing on the wall late – very late – and gave up the idea of a professional career in Engineering in 1971. I began teaching Remedial English in a secondary school, for which I received the same pay, although unqualified and inexperienced, as I had been receiving as a design engineer with a State Scholarship in Mathematics, a Cambridge degree, many publications in the top journals, some impressive research and development achievements behind me, and twelve years of experience.

Recently, when I was asked to give Science and Technology career advice to the pupils at my son's school, Haberdashers', I told them that if they took up such a career they would look forward to being on half pay, and that Britain was getting out of high technology. I told them that if they were really keen they should make certain to study foreign languages and think in terms of a career abroad.

Contrary to some reports, I have never attacked Finniston or his Committee of Inquiry into the Engineering Profession. However, I would now like to say that in my opinion there is no chance that their final report will be helpful to those who work in electronics and computer design. This is because none of the seventeen members of his committee is drawn from electronics, although electronics is a very large part of the engineering profession. (Computers alone are the third biggest industry in the USA.) The Committee will repeat the lie that our best engineers must be coaxed into rolling up their sleeves (like the good plumbers they really are) and getting into production. Because of their limited background, their lack of experience in high technology industry, the committee

members will not know that Production is a facet which tends to disappear as the sophistication of the technology increases. It is difficult even to find the production department in a high technology company, for instance one involved in advanced radar. If you think that Production is the essence of advanced engineering, you ignore the kind of message that Marshall McLuhan was trying to put across thirteen years ago. Notice that all examples on tv lauding the supposedly noble, against-the-tide first class engineer working in production are taken from old, declining industries. Such stories are merely another attack on high technology and its massive potential.

I am pretty certain that I can guess, from the tone of his letter, which company the fifty-one year old is working for. I would advise him that although the harassment of professional engineers is severe throughout British industry, in the case of his company it is a little worse than usual. The managing director of his company is systematically rooting out the engineering competence in his company, and by now has to a large extent completed the task. In five years' time, this managing director will be pilloried for destroying his company. But that is all in the future, and cold comfort for the fifty one year old.

*Ivor Catt
St. Albans
Herts.*

TALKING TO COMPUTERS

I was amused to read Mixer's remarks on computer programming (p.92, June).

The real reason we don't programme computers in the Queen's English is the same reason we don't write algebra textbooks in English prose – a formalized language is far clearer than a human language. There are things I can tell a computer succinctly in PL/I which I couldn't tell it (unless very awkwardly) in English.

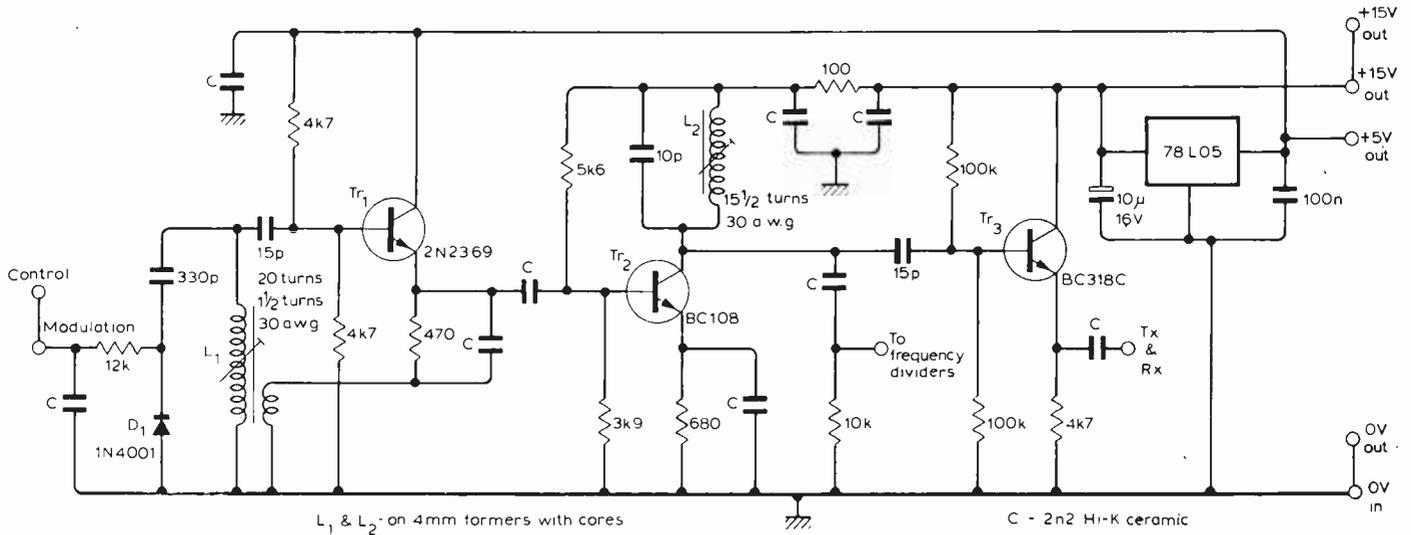
Human languages are context-dependent; to understand a human language the hearer needs much more background knowledge and must make many more assumptions than are necessary in understanding a formal language. Though computer handling of natural language is not far away and will be quite useful, formal programming languages will hardly be superseded.

*Michael A. Covington
Athens
Georgia, USA*

F.M. TRANSCEIVER SYNTHESIZER

Like Mr Hankins (Letters, June) I have constructed the synthesiser part of T. D. Forrester's two-metre f.m. transceiver (November and December, 1977). I have also found modifications essential to make it work. These are as follows:

1. L_1 was increased to 20 turns, with a separate 1-turn coupling coil. The varicap diode D_{17} was replaced by a device of greater capacitance – in my case a 1N4001 rectifier diode worked well – and C_7 was increased to 330 pF.



The modified v.c.o.

2. I added a level shifter between the t.t.l. and c.m.o.s. to provide a proper voltage swing - this was similar in design to Mr Hankins' circuit.

3. I found that the sensitivity of the 74LS74 was improved by increasing rather than decreasing R_{39} : I used a value of 10k Ω . The reason for this is that this allows the t.t.l. output to bias itself up to the logic "1" level; the input resistor forward biases the input diode of the gate to a point where its noise margin is effectively nil; and the full "gain" of the circuit is used to amplify the input signal.

As the input impedance of the low-power Schottky device is quite high - about 18k Ω - I found it made no difference to dispense with the buffer stage, Tr_4 , altogether and to connect the t.t.l. input straight through a 2n2 capacitor to the collector of Tr_2 . This saves 12mA or so of current.

4. In a further attempt to reduce the power consumption of the v.c.o. I removed the zener diode, D_{18} , and ran the oscillator stage (with changed resistor values) from the 5-volt t.t.l. power supply (obtained from a 78L05).

Replacing the 4049 hex c.m.o.s. buffer with a 4069 hex inverter might reduce the consumption of this chip, but unfortunately these devices are not pin-compatible.

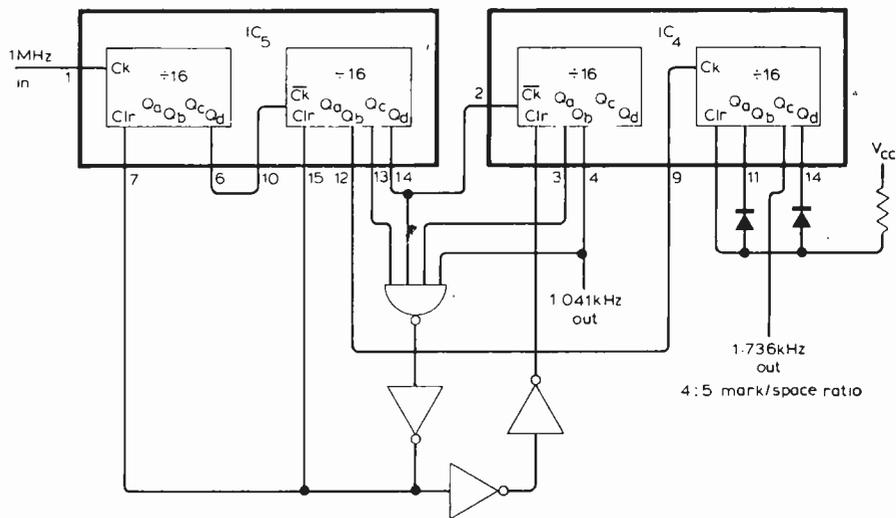
5. Apart from using a different arrangement of switches and gates to obtain the Tx, Rx, Tx(Rpt) and Rx(Rev. rpt) lines in the synthesiser logic, I also dispensed with many of the diodes and resistors: I used five diodes, and it can be done with four.

7. I used the spare divider section in IC_4 to obtain a frequency close to 1750Hz, for use as a toneburst generator.

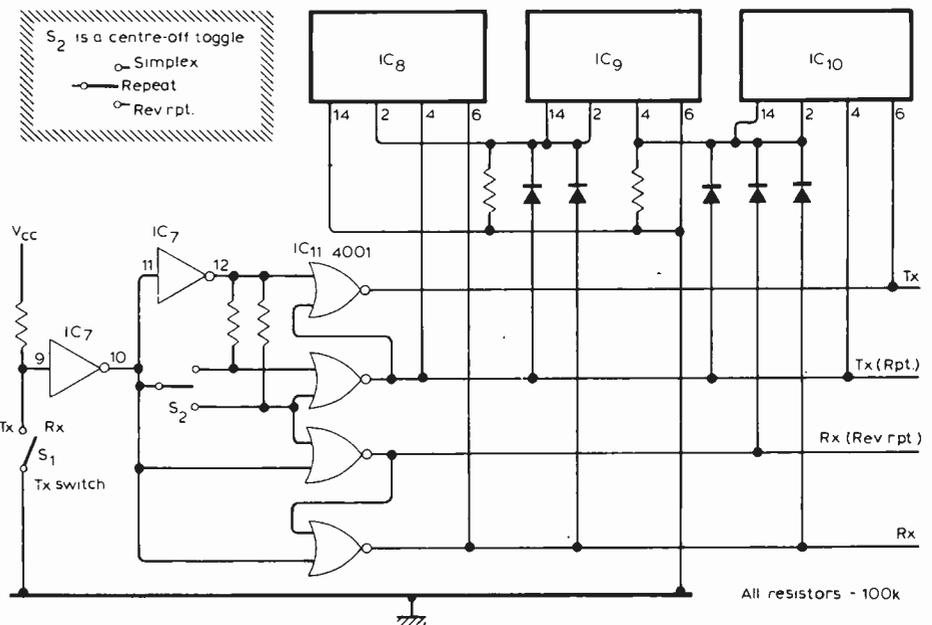
The accompanying circuits illustrate these points.

As a further suggestion the whole v.c.o. could possibly be replaced with a single t.t.l. v.c.o. chip, such as the 74LS324, which has a typical operating limit of 30MHz, a consumption of 30mA maximum, and, of course, a t.t.l. output. The output would be rich in odd order harmonics, so could be followed by a frequency tripler, with a doubler up to 144MHz: this is better practice as it avoids the generation of 48MHz.

A friend has observed that on his synthesiser (and on mine) the frequency overshoots its target value. This could indicate that the filter components after the phase detector have not been chosen for critical damping of the control loop. Another problem that has been suggested is in the quadrature detector



Repeater access tone generator. By applying a suitable positive-going pulse to pin 10 of IC_4 , the tone generator could be gated to provide a tone-burst.



The new Rx/Tx line generation and arrangement of diodes and resistors at inputs to b.c.d. adders.

of the i.f.: the use of a crystal here is really only suitable for ultra-narrow bandwidths, and a friend who has tried this circuit reports that it doesn't work.

I would appreciate advice and suggestions concerning this design from other constructors; for example, the use of alternative microphone pre-amp/compressor circuits instead of the rather expensive Plessey chip specified.

I would like to conclude by congratulating Mr Forrester on producing an alternative to the Japanese "black boxes", although I daresay some constructors will be less enthusiastic about the D.I.Y. (Design-it-Yourself) aspects than those of us with access to scopes, frequency meters, etc.

J. D. Stumbles

Imperial College Computer Centre
Exhibition Road
London, SW7

basis to meet the performance specification. A complete test schedule has been written and special production-test items have been made.

The second design is a simple two-op-amp thermistor-controlled R-C oscillator giving less than 0.001% total harmonic distortion from 100Hz to 10kHz — about 0.0003% at 1kHz — and less than 0.002% at 30Hz and 20kHz. The oscillator operates from a couple of PP9 batteries and takes about 14mA. In its present form the instrument is switch-tuned to give a number of spot frequencies throughout the audio band, but has a fine-tuning adjustment. A simple potentiometer output control is provided, this being adequate for most distortion-measuring applications.

The good performance of this latter oscillator results from (a) adopting an unconventional basic oscillator circuit giving far greater attenuation of thermistor distortion than does the usual Wien-bridge circuit, (b) giving careful thought to the theory of thermistor control-loop behaviour in order to obtain an optimum compromise between distortion and amplitude bounce, and (c) using good wide-band i.c. op amps. The total component cost, at Radiospares prices, including the instrument case, is about £14.

I would be glad to hear from any firm interested in the possibility of commercial exploitation of these designs.

Peter J. Baxandall
Malvern
Worcs.

AUDIO OSCILLATOR PRODUCTION DESIGNS

I have designed, and brought to a fully developed engineering state, two sine-wave audio oscillators having unusual and attractive performance characteristics.

The more complex design, which is neither a function generator nor a b.f.o., covers the complete audio spectrum in a single sweep, either by rotation of a dial having an accurately logarithmic scale shape, or by means of an internally or externally generated ramp voltage. Frequency-marker pulses are generated as the frequency sweeps through 0.1, 1 and 10kHz. A warble-tone facility is also incorporated, the oscillator being intended for acoustic measurement work.

The total harmonic distortion of the above instrument is less than 0.05% from 40Hz to 10kHz and less than 0.1% from 20Hz to 40kHz. The distortion is mainly second and third harmonic. The variation in output amplitude with frequency does not exceed ± 0.1 dB from 20Hz to 40kHz and there is no significant bounce.

The basic circuit is an R-C oscillator in which the R's have been replaced by transistor junctions whose d.c. bias voltage is varied to control the frequency in an inherently logarithmic manner. Two outputs in phase quadrature are available and could be used in association with a simple type of tracking filter. Very great care indeed has been taken to achieve accurate temperature compensation of frequency under production conditions.

In addition to the complete battery-operated instrument with the above facilities, ten of the main 11cm \times 12cm printed-circuit boards have by now been made, of which eight are in use by a large and well-known loudspeaker firm who have standardised on the design for their production test purposes. Experience with these eleven basic oscillators has shown that the design can readily be set up on a production

PROGRAMMING MICROPROCESSORS

Alas, K. G. Parr (August Letters) and I (June Letters) seem to be addressing different audiences, and so our viewpoints are difficult to reconcile. His admirable letter aligns with the opinions of at least one of my colleagues who is professionally engaged in systems development, both hardware and software, and who has both the opportunity (and perhaps the personality) to make effective use of microprocessors. I would be the first to concede that if one is committed to design these chips into, say, gambling machines (August issue, p.57) commercially, then one must try to be as impeccably disciplined as K. G. Parr suggests. Even so, as he tells us, "bugs still remain." And not only can they reveal themselves years later when an established programme faults, but so also one may discover errors in flow diagrams. This is not to disparage the flow diagram as a sometimes useful tool; but for a definitive final document I have usually found it more useful to edit a listing of the programme with some relevant textual comments, including reference to sources of algorithms, and with labelled arrows to indicate the purpose of loops, switches, jumps etc.

I hope that the readership of *Wireless World* includes already many "hobby programmers"; though surely all its readers are hobbyists at heart? I see that there will soon be a new generation of microprocessors designed to that they can be completely tested before use. So, hobbyists take heart if you are struggling with octal and hexadecimal. It can only be matter of time before decimalisation catches on.

Desmond Thackeray
Department of Chemical Physics
University of Surrey

DISCRIMINATIVE METAL DETECTOR

I read the article "Discriminative metal detector" by R. C. V. Macario (July issue) with interest. Although the author states that eddy currents mask diamagnetic effects in non-ferromagnetic samples it is worth noting that these eddy currents do cause an increase in oscillator frequency. This is, of course, in the opposite direction to that caused by ferromagnetic samples, where the ferromagnetic effect usually masks the eddy currents.

The effect of eddy currents can easily be demonstrated by comparing the influence of a closed loop of thin wire with that of the same loop opened. [Surely a coupled shorted-turn. — Ed.]

The directional information could be restored with headphones by using the Q outputs of IC₂ to drive a simple resistor ladder and voltage controlled oscillator (eliminating IC₃, etc.).

M. Walne
Halifax

More letters on the metal detector will be published next month. — Ed.

PICKUP-ARM PROBLEMS

I would like to point out that adoption of the technique proposed by F. Holloway (Letters, July) of producing discs by means of a cutting head on a radius-arm does not overcome the other serious shortcomings of the radius-arm reproducer so well highlighted in Mr Randhawa's article, i.e. side-thrust compensation and lateral balance.

There is no case for "concentrating the difficulty and cost of mechanical design once and for all in the recording equipment" as your editorial note suggests when parallel-tracking arm designs with electronic controls are now coming to the fore and which solve all three problems at a stroke.

Is it not about time that electronic engineers turned their attention from designing amplifiers with specifications of unnecessary excellence and concentrated their talents on programme sources such as tape and disc, which are still producing distortions that are audible?

R. Cooper
Sutton Coldfield

NEW PRODUCTS

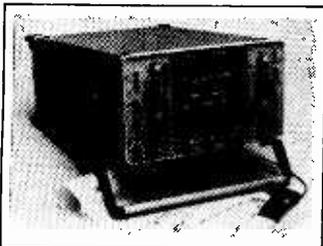
Filter

A bucket-brigade filter, said to be the first charge transfer device filter i.e., is introduced by the American Reticon Corporation. The circuits consist of a 64-stage, split-electrode brigade, the capacitors being the basic metal-oxide semiconductors. Low-pass and band-pass filters are available, each being tuned by varying the clock frequency, and possess a linear phase characteristic. Extremely steep rolloffs of over 200dB/octave are exhibited. An example quoted is that of a band-pass filter, in which the centre frequency is at 0.25 the clock frequency (250Hz-1MHz), the bandwidth 0.055 the clock frequency and the ratio between stop band and pass band attenuation is 52dB. The rate of change is 270dB/octave. Herbert Sigma Ltd, Spring Road, Letchworth, Herts.

WW301

Storage oscilloscope

Signal storage is performed digitally in the Advance OS4100 oscilloscope. It is a two-channel instrument using a digitizer and 1kbit r.a.m. to provide a bandwidth of 600kHz. An X-Y display



is also provided; the sum or difference of the two inputs can be displayed and an unusual feature is a triggering window, in which triggering takes place outside two thresholds. Triggering can be obtained from 2mm of trace deflection and a delay can be used to display events occurring up to a quarter of the time-base period before the trigger. The trace consists of a number of dots, which can be 'smoothed' if required, and a split-trace mode can be selected where alternate samples can be sorted and viewed against a new trace for comparison. Gould Instruments Division, Roebuck Road, Hainault, Essex IG6 3UE.

WW302

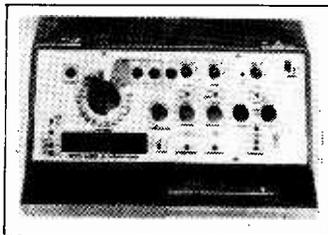
Electronic wattmeter

An electronic multiplying technique used in Feedback Instruments EW604 power meter allows wide frequency and power ranges to be handled by the one instrument. It also means that waveforms other than sinusoids can be measured. The range of the instrument is 0.25W to 10kW f.s.d., 5V to 1kV r.m.s., 80mA to 10A r.m.s., all in the frequency band from zero to 20kHz. To reduce effects on the circuit under test, the input resistance is 5kΩ per volt, drawing 200μA full scale, and the resistance offered to a current input is 60mΩ. Feedback Instruments Ltd, Park Road, Crowborough, Sussex TN6 2QR.

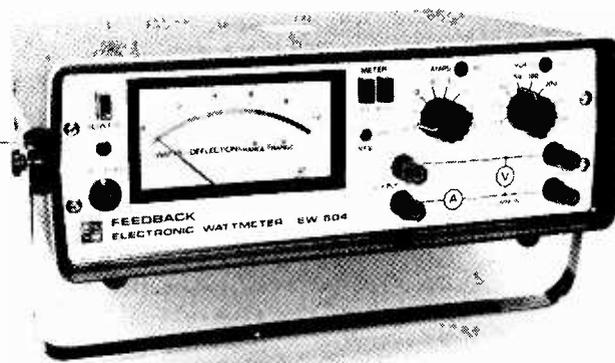
WW303

Digital tester

Comprehensive facilities for measurement in the time and frequency domain, voltage, resistance and temperature are contained in one new Tektronix instrument, the 851 Digital Tester. A 5-digit display is fed by a 35MHz counter, which performs all the usual frequency,



time, ratio and event counting from three input channels; a digital multimeter of 4½-digit resolution measuring voltage, resistance, temperature and the thresholds of the input channels, together with polarity. The in-

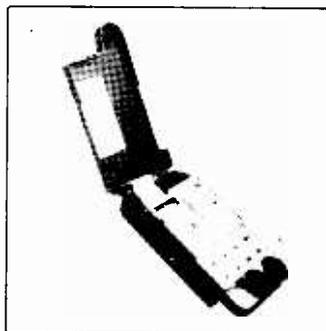


puts can be set to t.t.l. as a calibrated knob setting or to any thresholds compatible with other logic families within ±30V. All necessary probes are housed within the case of the instrument. Tektronix UK Ltd, Beaverton House, P.O. Box 69, Harpenden, Herts.

WW304

Audio system measurement

ATR-1 Audiotracer consists of a voltage-controlled oscillator, logarithmic in frequency from 20Hz-20kHz or 200Hz-200kHz, an

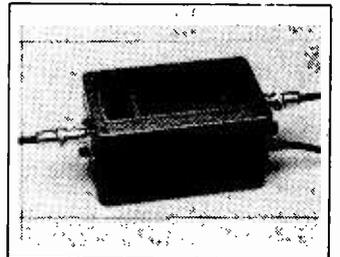


output amplifier, a log. or linear input amplifier, a true r.m.s. rectifier and a pen recorder. The whole thing is contained in a 11 × 5½ × 3in case and is intended to measure and record the frequency/amplitude performance of electronic or electro-acoustic systems. A measuring microphone and an artificial ear coupler are available for acoustic measurements. The unit is mains-powered. ATR-1 is made by Neutrik A.G. of Switzerland and distributed here by Eardley Electronics Ltd, Eardley House, 182/4 Campden Hill Road, Kensington, London W8 7AS.

WW305

Frequency scalars

Frequency scalars are dividers, used to reduce high frequencies to within the capability of low-frequency counter-type frequency meters. There is no display, and the division factor must be re-applied to the counter display. Two units produced by MTG fulfil this function, the PS1200 enabling a 10MHz counter to measure a frequency of 1GHZ (PS520 up to 500MHz). Both units will accept 10mV minimum input and are well protected against overload and out-



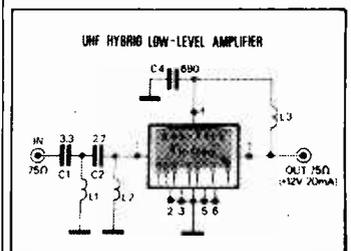
put short circuits. Versions offering division ratios of 10 or 100:1 are available. MTG (Instruments) Ltd, Beacon House, Christchurch Road, Lansdowne, Bournemouth, Dorset BH1 3LB.

WW306

U.h.f. amplifier

A two-stage hybrid amplifier, the SH120A offers 16dB of gain, which is said to be constant from 40-900MHz, and a noise figure of 5dB. It is a low-level aerial amplifier, being terminated in 75 ohms in and out and its maximum output is 100mV. Voltage supply needed is 12V. SGS-Ates (UK) Ltd, Planar House, Walton Street, Aylesbury, Bucks HP21 7QN.

WW307



Pressure transducer

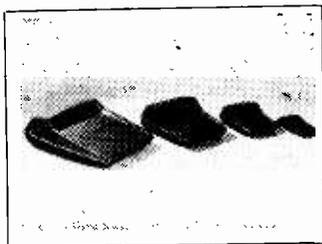
Two transducers from Philips convert pneumatic pressure to standard 4-20mA signals for transmission or processing. A metric type (PR9363/20) covers the range 0.2-1 Bar, while an Imperial version (PR9363/30) works from 3-15 p.s.i., both units conforming to DIN 19231 and IEC 381. Unstabilized power can be used. The principle of the devices is the deflection of a strain-sensitive diaphragm by the pressure, unbalance in the strain elements giving rise to an output voltage. Pye Ether Ltd, Caxton Way, Stevenage, Herts.

WW308



Instrument cases

The new 'Princess' range of cabinets from West Hyde is intended to house the various kinds of equipment to do with computers, e.d.p. and calculators. The cases are made from ABS, in two horizontal parts, and have two frontal surfaces, at differing slopes, for mounting controls and displays. The plastic can be drilled and cut and is finished in an imitation leather texture, in black



or black and tan. The internal dimensions are keyed to the Eurocard shape, the largest case taking double Eurocards. West Hyde Developments Ltd, Unit 9, Park Street Industrial Estate, Aylesbury, Bucks HP20 1ET.

WW309

Frequency meter

A u.h.f. counter, of the non-displaying scaler type, of high stability and accuracy offered by the American firm of Davis is the 7208 Frequency Counter. The instrument will measure frequency up to 600MHz at 10mV

up to 60MHz and 100mV at 600MHz—less with an optional preamplifier. Crystal frequency error is 1 part per million or 0.5 p.p.m. with the optional oven, while the drift is 1 p.p.m. per hour or 0.5 p.p.m. per month, again with the oven. An l.e.d. display of eight digits is used, with decimal point being adjusted by the switching. Davis Electronics, 636 Sheridan Drive, Tonawanda, New York.

WW310

Spray etcher

This unit, from P.B. Electronics, is a double-sided etcher, which is small enough to stand on a bench and fast enough to etch a 12in square board in less than three minutes. It uses four gallons of etchant, which is kept at 120°F, and a timer stops the pump after a given time. A lid interlock reduces the chances of etching the ceiling. P.B. say that the etcher will be followed by other units for sensitizing, exposing and developing. P.B. Electronics (Scotland) Ltd, 9 Radwinter Road, Saffron Walden, Essex CB11 3HU.

WW311



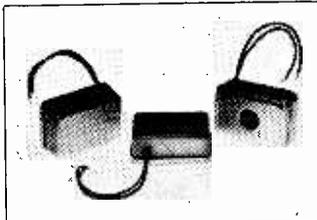
Voltage-to-frequency converter

The option of linear or logarithmic conversion and a dynamic range of more than seven decades are the features of the Aragorn VFD1 module. The output is a 10V square wave, which responds to a 0-90% step within 1 cycle, the maximum output frequency being 2MHz or 20kHz to order. Accuracy of conversion is 0.1% and frequency stability is 200 parts per million at 1kHz. Voltage supply needed is 15-0-15V and the module measures 45 x 30 x 16mm, with pins on a 0.1in grid. Aragorn Dynamics Ltd, 8 South Side, Clapham Common, London SW4 7AA.

WW312

Buzzers

These are miniature, electronic units for use in clocks, timers, telephones, etc. They are for use where 2.5V, 6V or 12V supplies, of low stability, are present. At least



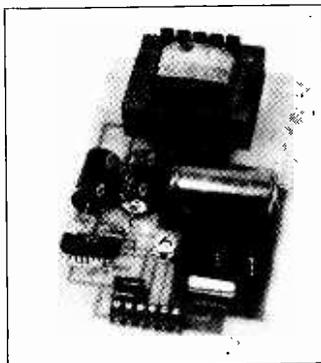
70dB relative to 0.0002 dynes/sq.cm. is produced 22cm away from the unit, while taking around 20mA from the supply. Type number is GA100. Highland Electronics Ltd, Highland House, 8 Old Steine, Brighton, East Sussex BN1 1EJ.

WW313

V.m.o.s. memory

Two static 4K memories from AMI use the v.m.o.s. technique to provide high-speed access and high density. The S2114 is organized as a 1024 x 4 bit r.a.m. and is intended as a high-speed Intel 2114 replacement, for which it is claimed to be pin-compatible. Maximum response time can be down to 150ns. The S4017 is a 4096 x 1 bit r.a.m. with a response time down to 55ns. Both types are usable with t.t.l., operating from 5V, and are contained in 18-pin plastic or ceramic packages. AMI Microsystems Ltd, 108A Commercial Road, Swindon, Wiltshire.

WW314



P.s.u. for logic and linear

Where 5V digital circuitry and linear i.cs are to be used together the power needed can conveniently be provided by the Lascar 3-rail power supply. Outputs are 5V at 1A and tracking rails of 5 to 15V, positive and negative, at 100mA. The single and dual supplies are isolated from each other and both are well protected. A 160 x 100mm board carries the whole unit. Lascar Electronics Ltd, P.O. Box 12, Module House, Billericay, Essex CM12 9QA.

WW315

Signal conditioner

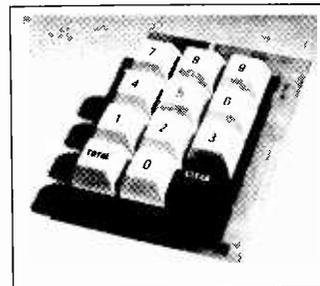
Platinum resistance elements are widely used in temperature measurement bridges, but their non-linearity restricts them to laboratory use, in the main. Ancom have now produced a signal conditioner, 15RP-3, to allow the use of a platinum sensor with a linear measuring system — a digital panel meter, for example. The 51 x 29 x 16mm, encapsulated, p.c.-mounted module provides 10mV per degree Centigrade output, within the -110°C to 100°C working range, with errors of better than ±0.1°C and ±0.1% of reading. Power needed is 15-0-15V at 10mA (each rail).

Ancom Ltd, Devonshire Street, Cheltenham, Glos. GL50 3TL.

WW316

Keypad

This 12-key pad from FR Electronics is of low height (less than 25mm when the buttons are down — unspecified when they are not) and is said to offer a high degree of reliability as a consequence of its use of reed switches. The keys are mounted on a printed board, with



an edge connexion, from which the outputs are in binary form, compatible with t.t.l. levels. Voltage supply needed is 5V and two-key rollover is a standard provision. F.R. Electronics Ltd, Wimborne, Dorset BH21 2BJ.

WW317

Microwave f.e.t.

A noise figure of 1.7dB or better at 4GHz and a useful range of application from 1-12GHz is claimed for the HFET-1102 GaAs f.e.t. from Hewlett-Packard. Minimum associated small-signal gain is quoted as 11dB at 4GHz. The encapsulation is HPAC-100A. Hewlett-Packard Ltd, King Street Lane, Winnersh, Wokingham, Berks RG11 5AR.

WW318

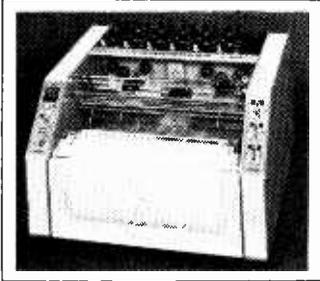
Coaxial relays

Between 2 and 30 coaxial inputs, switching to a single coax. output, all of either 50 or 75-ohms, form a single-pole, multi-way, coaxial r.f. switch, with isolation of around 100dB. Unselected inputs may be earthed, open-circuited or terminated in the relay block; contacts of the reed

relay can be expected to endure up to 20 million operations and will handle 250mA at 100V or less, depending on the switching mode. The switching signal can be anything from 5 to 50V d.c. or a.c. Hercoax Ltd, Plumpton House, Plumpton Road, Hoddesdon, Herts.
WW319

Chart recorder

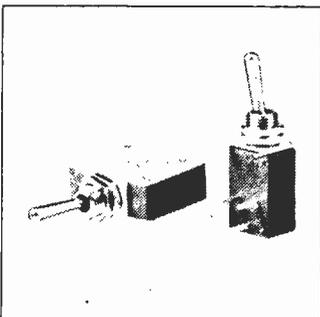
Unicorders are desk-top, potentiometric chart recorders with three to six steel pens, which may overlap. A complete traverse of



the 250mm chart takes 0.3s, and the pens can be lifted individually, by solenoid as an option. The 20m chart is stepper-motor driven at a speed selected from 199 dial-selected or remotely-controlled possibilities. Plug-in amplifiers accommodate the voltage range 1mV to 200V, currents from 1µA to 500mA and thermocouples at 0°C-1600°C. Common-mode rejection ratio is 170dB at zero frequency and 160dB with alternating signals. Rostol Ltd, 33 Byron Road, Earley, Reading, Berks IG6 1EP.
WW320

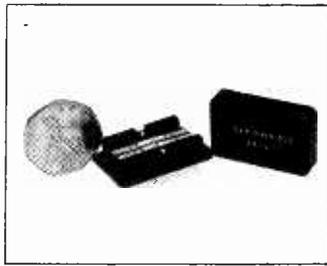
P.c. switches

APEM 21000N toggle switches are designed to be mounted on printed boards, projecting only 9mm above the surface. Single-pole switches are 11 × 7 × 7.8mm; double-pole 21 × 17 × 7.8mm, excluding pins. They possess either two or three positions, latched or momentary and have silver or gold-plated contacts. Voltage and current capability for the silver contacts are 250V a.c. and 2A a.c. and the gold contacts will control 10mA at 50mV. Iskra Ltd, Redlands, Coulsdon, Surrey CR3 2HT.
WW321



Mechanical filters

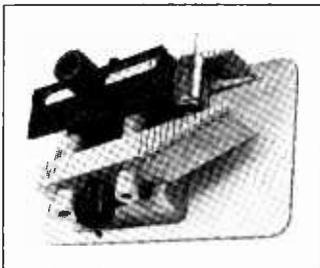
Low-frequency filters using flexure-mode mechanical resonators are announced by Rockwell-Collins. Nickel-iron bars and piezoelectric ceramics are the active elements, pro-



ducing a response of 0.2 to 1.5% bandwidth/centre frequency which is equivalent to a Q of 500 to 70. The filters work in the range 3.5-70kHz, with an insertion loss of 2.5dB, and terminating resistance of 33kΩ. Centre frequency varies less than 4Hz over a 0°C to 60°C range of temperature. G. A. Stanley Palmer Ltd, Elmbridge Works, Island Farm Avenue, West Molesey Trading Estate, Surrey KT8 0UR.
WW322

Lead forming tool

Consistency and accuracy in the bending of component leads for printed-board mounting is the aim of the Litesold Opsec tool.



Components are dropped against a graduated step into the slots on the jaws, which are also graduated to take components of varying lengths. The tool is made of high-impact plastic and has non-slip feet, which can be screwed to a bench. Light Soldering Developments Ltd, 97/99 Gloucester Road, Croydon, Surrey CRO 2DN.
WW323

Low-resistance meter

A main-frame and plug-in adaptors, the 1700 series, from Electro Scientific Industries are capable of measurement and numerical display of resistance values in the range 1 micro-ohm to 200 milliohms, in which lie the resistances of contacts, inductor windings and printed-board

tracks. One of the plug-in units offers the option of d.c. measurement for reactive components or a pulsed test for those components whose characteristic would change if current were to be passed continuously. Tranchant Electronics Ltd, Tranchant House, 100A High Street, Hampton, Middlesex TW12 2ST.
WW324

R.f calorimeter

A numerical reading in kW, allied to the calorimetric determination of radio-frequency power, enables convenient power measurements to be made by the Bird 6080 up to 80kW with a maxi-



mum error of ±3%. Temperature sensors and water flow monitors are remote, being connected via an 8ft cable to the power meter. Aspen Electronics Ltd, 2 Kildare Close, Eastcote, Middlesex HA4 9UR.
WW325

Dual-lamp switch

A dual-lamp switch, available from Pye Electro-Devices Ltd, has a ¾in-square cap incorporating two T13/4 lamps with a midget flange base and a matching indicator. The cap has horizontal or vertical split screens, with five choices of filter and eight lens colours for each of the two zones of the screen. Silver or gold contacts, with momentary or alternative actions, are available in s.p.d.t. and d.p.d.t. configurations. Pye Electro-Devices Limited, Controls Division, Exning Road, Newmarket, Suffolk CB8 0AX.
WW326

Darlington's for tv deflectors

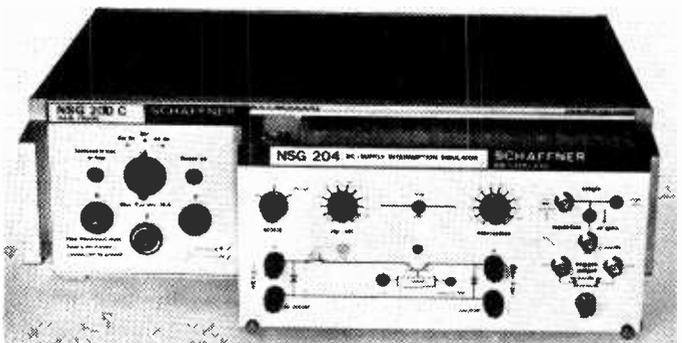
Two high power transistors, the BU806 and the BU807, are particularly suitable for use as horizontal deflectors in black and white televisions. The low driving power typical of the Darlington configuration allows elimination of the driving transformer and consequently reduces the cost of the stage. The high switching speed required is guaranteed by the presence of an integrated diode which extracts the charges accumulated during conduction. A clamper diode is also integrated in the devices. These devices, when used with the integrated horizontal deflection circuit, the TDA1180, no longer require a driving transistor and consequently the cost of the system is further reduced. SGS-Ates (United Kingdom) Limited, Planar House, Walton Street, Aylesbury, Bucks, HP21 7QN.
WW327

Fibre light guides

General purpose fibre-optic light guides in a p.v.c. sheath and terminated in stainless steel end fittings are produced by Valtec. Fibres of 0.002 in diameter are used in single, bifurcated and trifurcated branch types in lengths up to 30 ft attenuation is 600-700dB/km, numerical aperture is 0.56 with a 68 degree acceptance angle. Fiberoptics Division, Valtec Corporation, West Boylston, MA, USA 01583.
WW328

Supply failure simulator

Interference simulation by the Schaffner NSG200 series is extended to allow the simulation of d.c. power failure and interruptions. A new plug-in unit, the NSG204, is for use with the 200C mainframe, and permits d.c. supplies between 5V and 220V at up to 10A to be interrupted for adjustable duration between 1ms and 2s. The interruptions may be generated by an external trigger, a push-button or an internal timing generator, working between 5Hz and 0.1Hz. Switching time is around 2µs off and 1µs on. Lyons Instrument Ltd, Hoddesdon, Herts.
WW329



π-test

Embarrassing moments experienced by television and film "personalities" can't be the same kind as those I have — they always seem so delighted to tell anyone who asks them all about the grisly faux pas and face-burning ineptitudes they've committed. I can't bring myself to tell anyone about mine, but I'm always happy to hear of other people's. The only time I've ever been able to see one of my own episodes as even remotely funny was when I sidled up to my wife, who was trying on a hat in a shop and said "You can always use it as a flower-pot when you're tired of it" and yes, you've guessed it, it wasn't my wife. The woman wasn't trying on hats, either; it was her own.

"What's all this to do with electronics?" I hear all my readers ask, both at once. Well, it's just that we were talking, this morning, about Chairman's Visit time, from which many of us have suffered. You know what I mean; the one day a year that makes it all worthwhile. White lines are painted on the factory floor, guards are put on machines and all design engineers are told to wear their clean shirt and tie and have a shave if they can possibly manage it.

Our chairman used to come round the labs. and demonstrate that democracy was still a force to be reckoned with by speaking to us. Actually, he usually directed his questions to us by way of the chief engineer, and didn't bother listening to the answers, but the intention was there. We always had to show him the newest bits of gear we had prised out the buying officer during the year so that the management could show him how forward-looking they were and, this year, we had bought an oven for environmental testing of components. It was a big metal box, with lots of terminals and leads sprouting from them and any amount of meters and generators clustered round it. The chap who was using it explained what it was for but seemed a bit reluctant to



open the door. It would upset the test run and probably spoil a lot of hard work and other protestations, but the Man wanted a look inside, so with a dramatic flourish, the door was opened. There were three shelves inside and nothing else at all, except that right in the middle of the centre shelf was a very small pork pie, gently steaming.

Not a word was spoken. The group moved on and the only sign of anything amiss was that our chief engineer was abstractedly chewing his tie. I don't remember what happened to the chap with the pie — he's probably a permanent lab. assistant now.

Timeo Danaos

It seems we're about to be saved. You can all come back in off the window ledge, because Whitehall has decided that the electronics industry could do with a hand and is intent on injecting £50m or so to revitalize, rejuvenate and rekindle the sparkle in our eyes.

Well, that's great, but you will, I hope, forgive me if I don't instantly leap to my feet and turn cartwheels all the way along Stamford Street. I'm too old and frail, for one thing, and the other is that I have this sudden presentiment of

doom. I put it down to the SADIM syndrome, so named because of a Greek character — a king, as it happens — who had a regrettable tendency to turn everything he touched into clay; he could have made a fortune in the china industry but he had no marketing sense. There he sat all day, strumming on his bouzouki and surrounded by little piles of pure gold; every now and then he would snatch a pile of gold and, calling on his training as an alchemist, transmogrify it into base clay. (His old prof. used to say that the lad had never seemed to get the point of the subject).

The striking thing about all this is, though, that not only did we learn our democracy from chaps like that, but some of his financial expertise seems to have rubbed off too. The cream of our society at Westminster have, it seems, only to take a passing interest in an industry for it to become a disaster area. There is no need to plod wearily through the list of victims, but if you kick off with the Brabazon and doff a mental hat at the TSR2 and the Hovertrain, finishing up with British Leyland (or whatever they call it these days) and Strathearn you'll see what I mean. It isn't a view of life to make one happy at the sight of Ministers bearing gifts.

Still, the china industry could have a rosy future.

Byter bit

Among the faults of David Bligh, excessive faith in I.s.i.

was very probably the worst, though in this field he was the first. He gathered chips from every source, attended every single course and ultimately he was known as Dai the Sums, and stood alone. As hardware goes, it came and went (around a thousand pounds he spent) but gradually, he amassed sufficient gear, and stopped his fast. Computers large, computers small, Dai knew the workings of them all. His own could indicate, at speed with all the confidence you need the contents of its ROM, in clear on a v.d.u., and bend your ear with cacaphonic sounds of bytes in battle for their storage rights. When Dai was asked if this was what his monster did, he waxed quite hot and instantly applied his mind to write a programme of a kind to show quite positively that computing's really where it's at. His friends all gathered round to stare being sure he'd finished with hot air. He ran the tape — the display flashed, the printer rattled — keys were bashed. Then, on the screen in letters twee was printed "Pawn to King's Knight Three".

The move was made, but all in vain, because a rook in wait had lain; it sidled gently up and said "Checkmate, old son, afraid you're dead!"



Uncompromising performance. Incredible price. A professional 3½ digit DMM Kit for less than £ 50.00

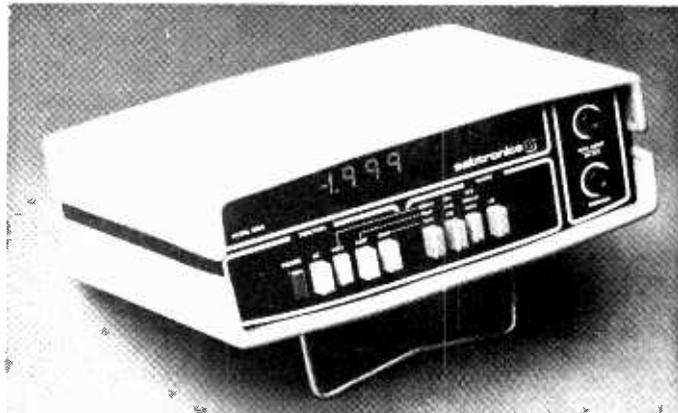


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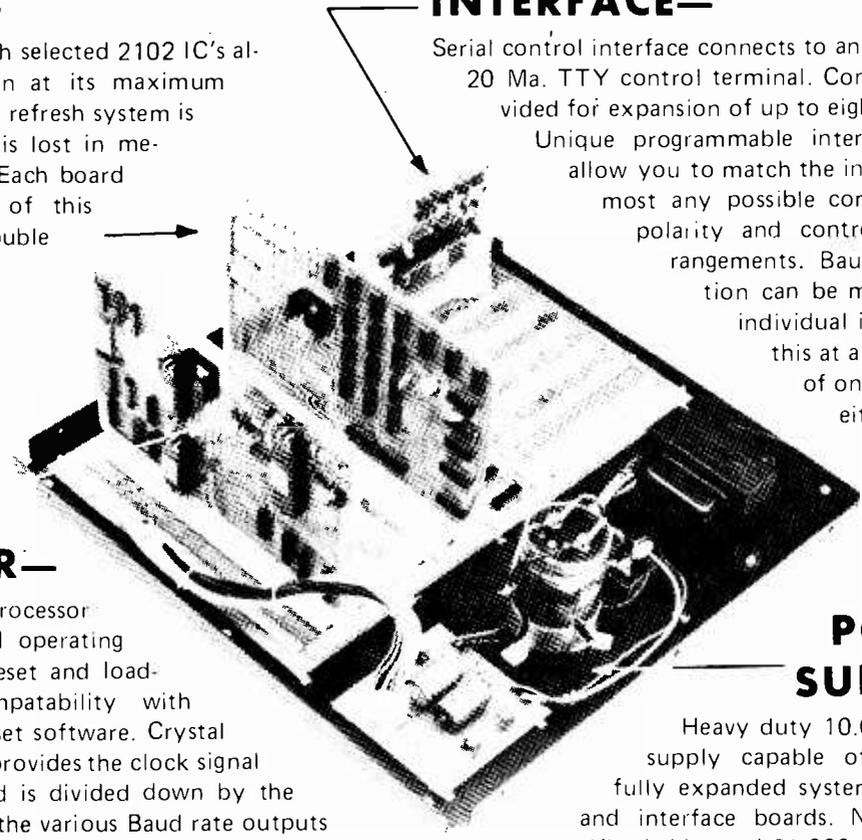
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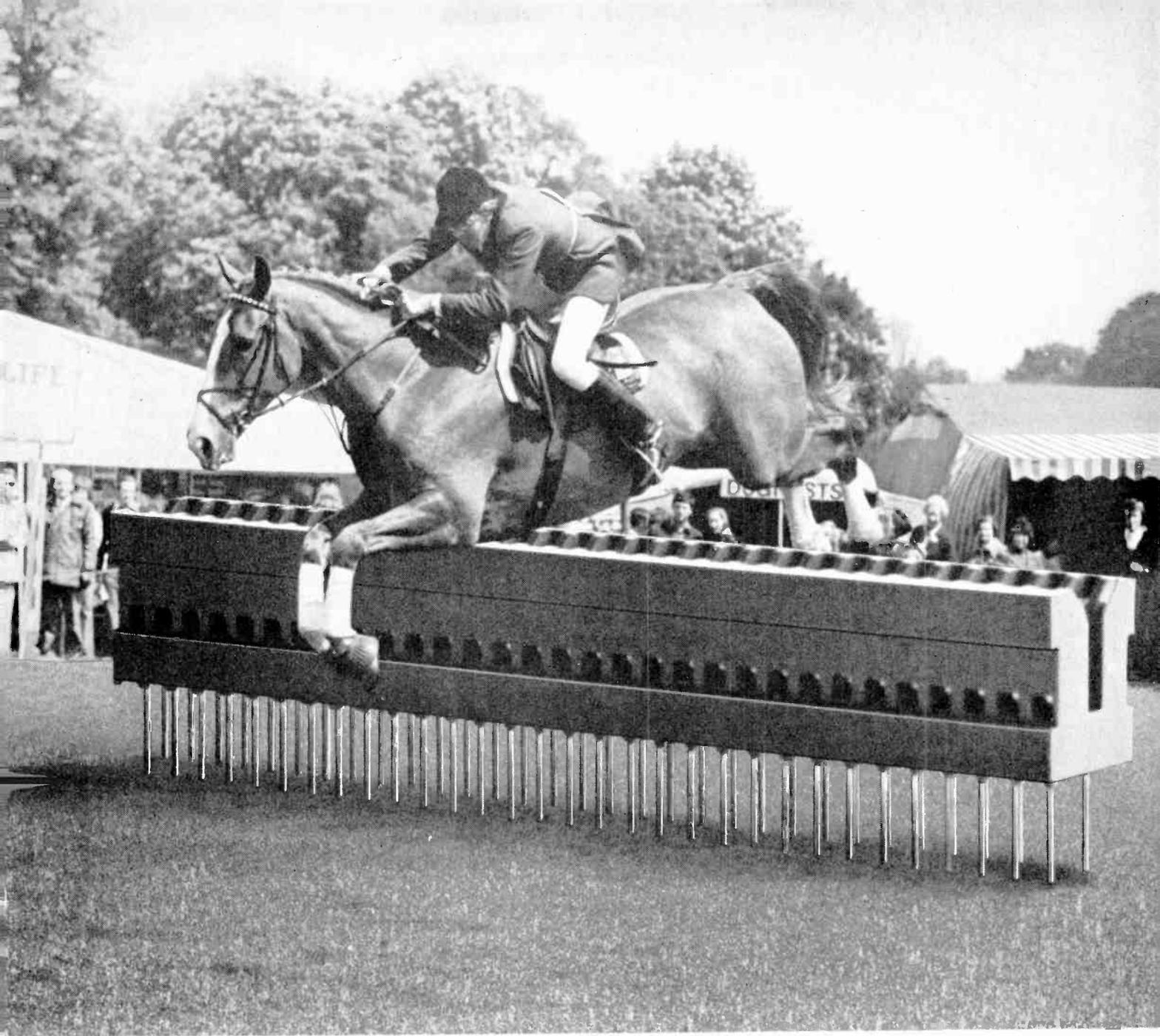
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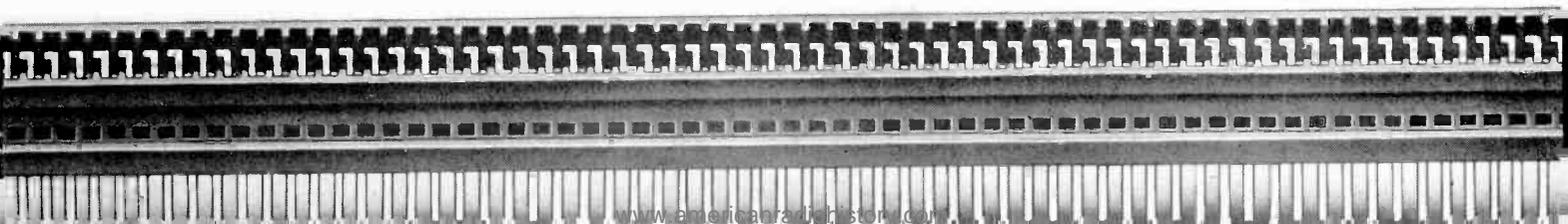
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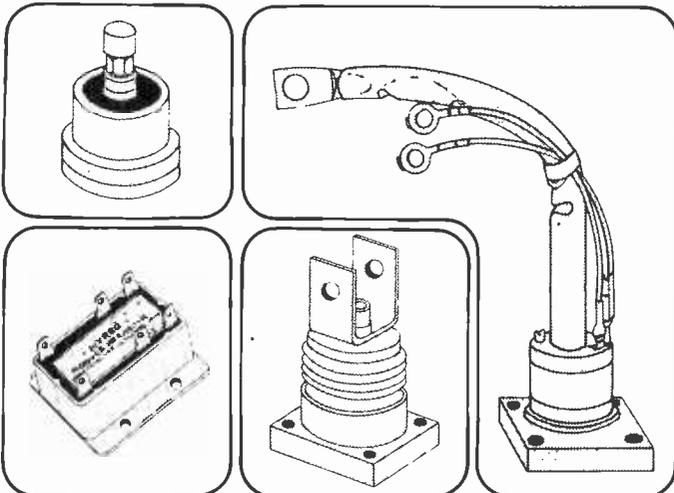
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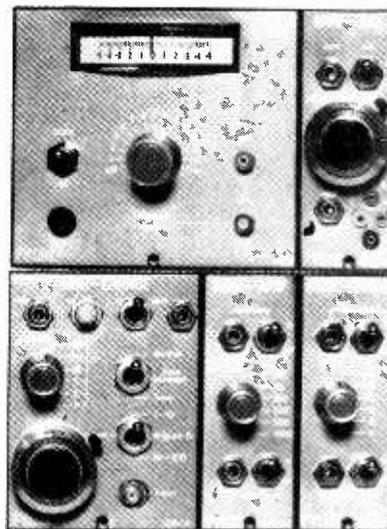
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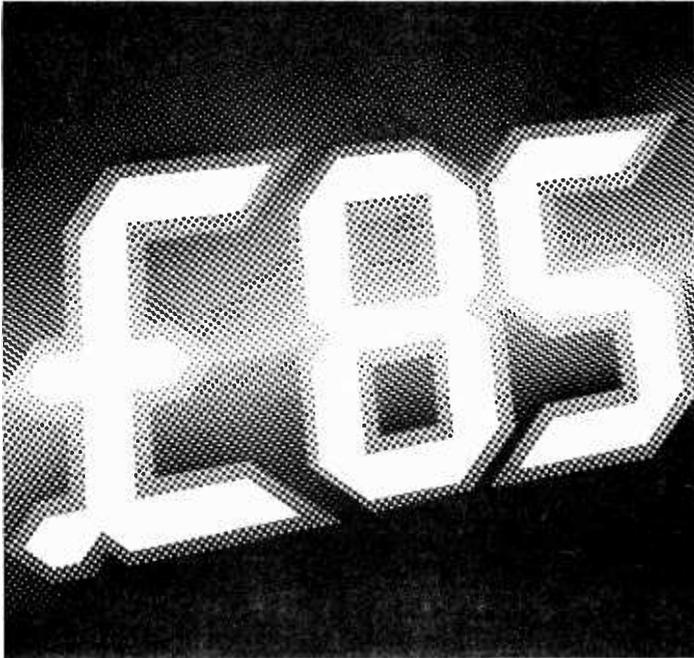
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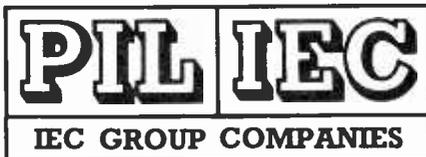
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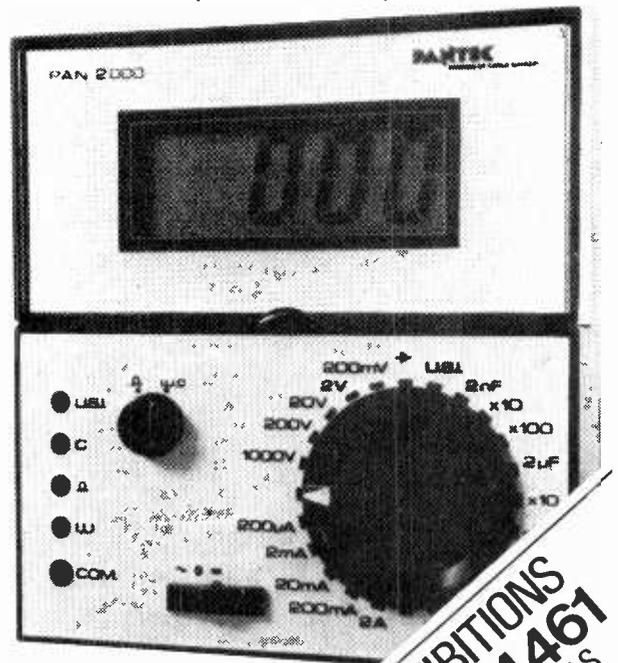


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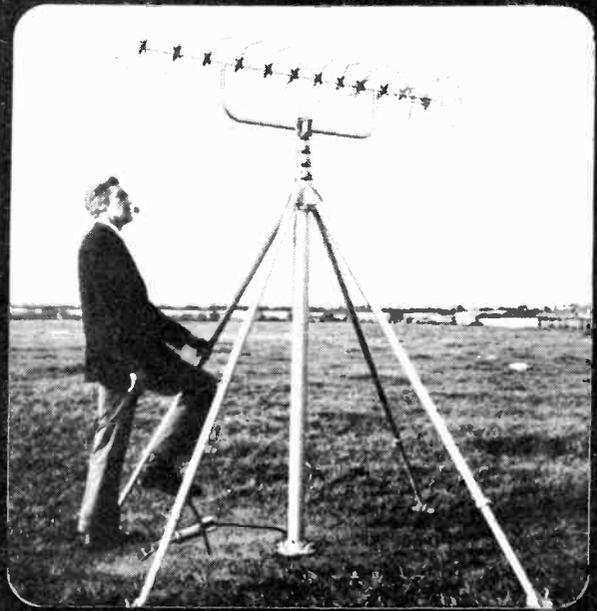
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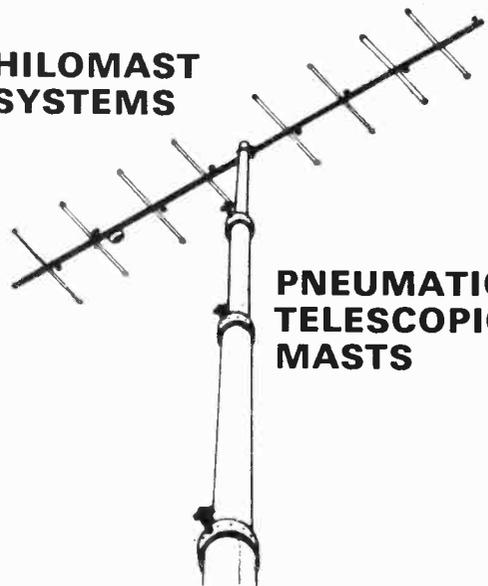
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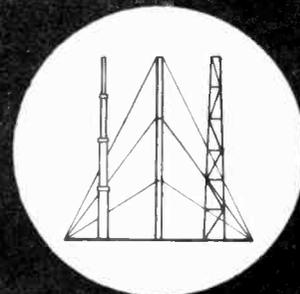
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7412	17	24	7482	69		74147	175		74200		74390	140	
7413	30	52	7483A	97	110	74148	109	105	74202	374	74393	140	
7414	51	130	7484	97		74150	99		74221		74395	110	
7415	30	24	7485	104	99	74151	64	84	74240	100	74396	133	
7416	30	24	7486	31	40	74153	64	54	74247	75	74398	200	
7417	30		7489	205		74154	96		74248	90	74399	150	
7420	16	24	7490	33	90	74155	54	110	74249	90	74445	92	
7421	29	24	7491	76	110	74156	80	110	74251	90	74447	90	
7422	24	24	7492	38	78	74157	67	55	74252	105	74490	140	
7423	27		7493	32	99	74158	210		74257	108	74658	110	
7425	27		7494	78		74159	210		74261	105	74670	220	
7426	36	27	7495A	65	99	74160	82	130	74266	40			
7427	27	29	7496	58	120	74161	92	78	74273	90			
7428	35		7497	85		74162	92	130	74275	175			
7430	17	24	74100	119		74163	92	78	74279	52			
7432	25	24	74104	63		74164	104	130	74283	120			
7433	40	32	74105	62		74165	105	110	74289	340			
7437	40	24	74107	32	38	74166	20		74290	90			
7438	33	24	74109	63	38	74167	20		74293	95			
7440	17	24	74110	54		74168	85	200	74295	120			
7441	74		74111	68		74169		200	74298	100			
7442	70	99	74112	88	38	74170	230	200	74299	300			
7443	115		74113	38	78	74172	625		74300	350			
7444	112		74114	38	78	74173	170		74302	350			
7445	94		74116	198		74174	87	120	74324	115			
7446	94		74118	83		74175	87	110	74325	140			
7447	82	99	74119	119		74176	75		74326	140			
7448	56	99	74120	115		74177	78		74327	95			
7449	17		74121	25	50	74178	85		74352	95			
7450	17	24	74123	48	55	74181	165	350	74353	100			
7451	17		74124	48	88	74182	160		74358	110			
7453	17		74125	38	44	74183	135	210	74260	26			

From the World's leading radio innovation source:

On this side of the page, we offer you the leading products from the world of wireless. We are continually reviewing and adding to our range, and this month we feature some of the more significant additions to our range. All available in depth, and usually ex-stock too!

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Ambit offers a very wide range of low cost meters, together with the unique 'Meter Made' scale system for professional grade scale customizing.

Series	Scale	Area	illumination	cost*
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920	30x50mm	from behind	275p	
930	36x63mm	internal	12v	375p
940	twin 35x45mm	from behind	350p	
950	55x45mm	from behind	300p	

Stock movement 200uA/750Ω. The 930 series is 5% linear, others are 77uA at 50% FSD. These and many others available in quantity for OEMs SAE for full scale details please. (Not in cat.)

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TBA651 LF/30MHz linear system	1.81
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filters in 5 groups	50p
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BLR3107N Stereo 4k7 impedance	215p
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BLR3152 Mono 4k7 impedance	100p
BLR3157 Mono 4k7/3k0 imp	100p

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MFL series 4.5/7kHz BW on 455kHz	195p
MFK series 7.9kHz BW on 455kHz	165p
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CF455B/6/8kHz min + 2IFTs	65p
CF470C 6kHz on 470kHz	60p

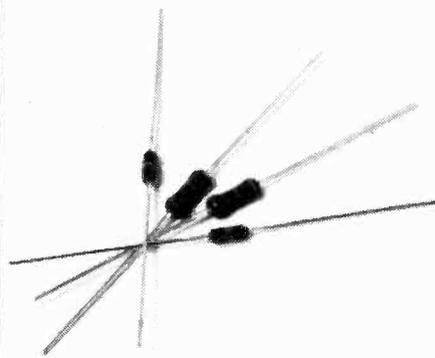
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Quadrature detectors for CA3089E etc

KAC5K586HM single	33p
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Treble-approved metal oxide resistors from ITT - of course



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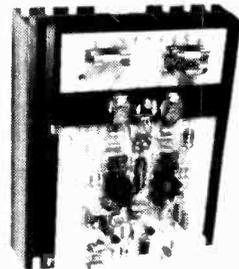
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WW-049 FOR FURTHER DETAILS

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WW-085 FOR FURTHER DETAILS

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MP-68	M6800 by Southwest Technical. With 4K RAM, Mikbug ROM and TTY/RS232 input output interface	275	330
CT-64	VDU controller with keyboard, serial ASCII	230	315
CT-VM	Video monitor for use with CT-64, 12 volt operation		140
AC-30	Freestanding dual Kansas-City cassette interface	60	100
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MP-8	8K memory board	170	195
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MOVING COIL PRE-AMP MC1

This stereo module uses multiple input transistors to achieve 65dB s/n. Sensitivity is switched 70 or 160μV for 3.5mV output

POWER SUPPLY

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POWER AMP KIT £32.40

PRE-AMPS:
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WW-122 FOR FURTHER DETAILS

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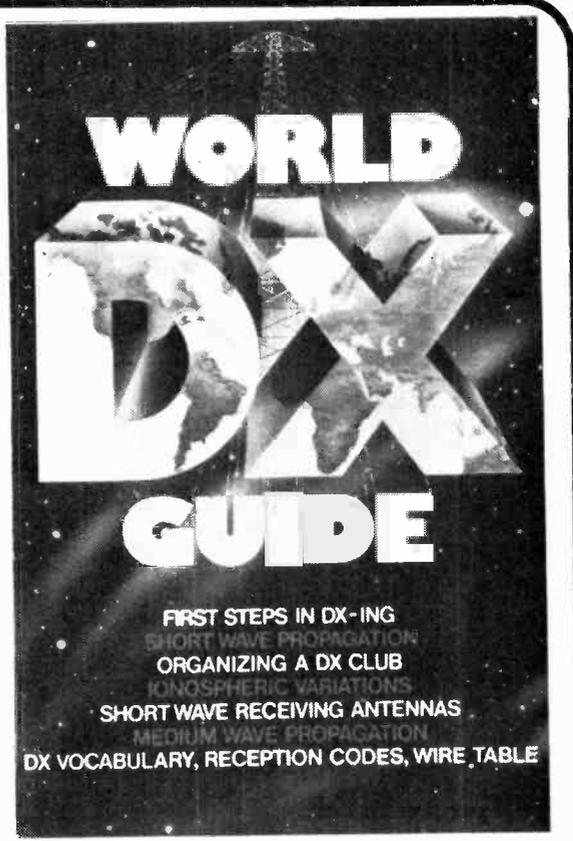
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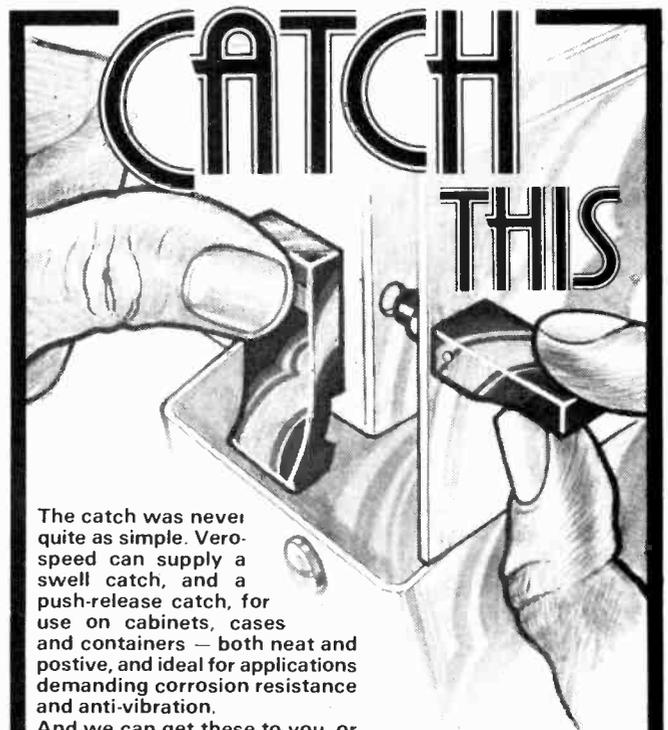
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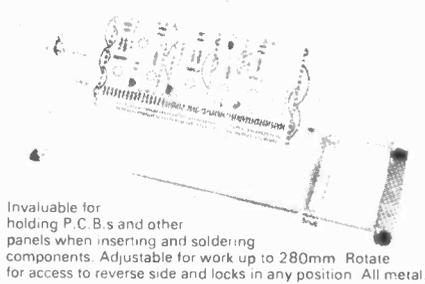
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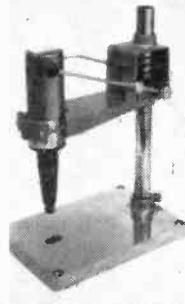
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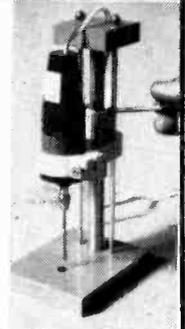


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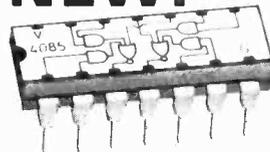
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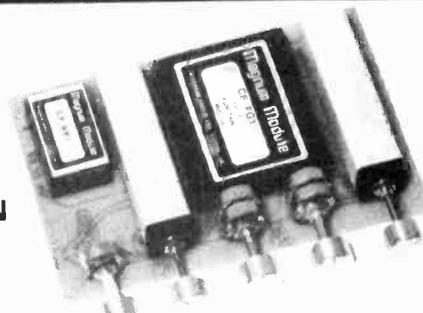
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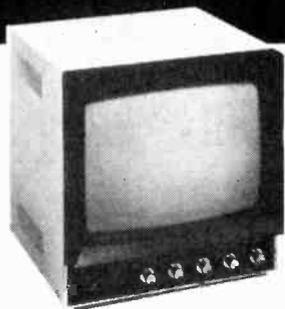
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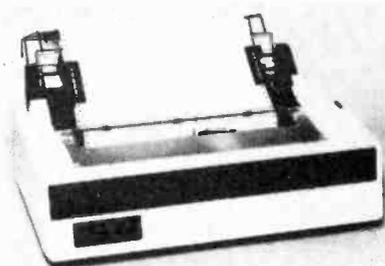
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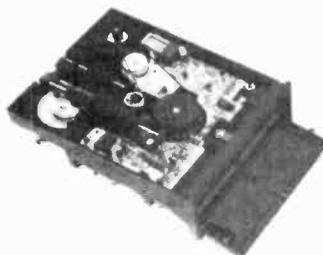


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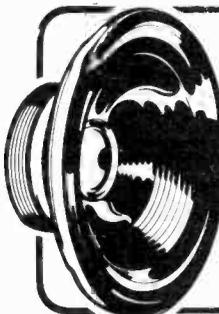


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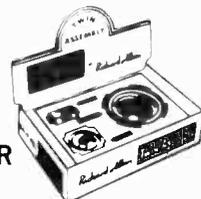


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Speaker Kits	£2.50 pair
Mag. design kits	£3.50 pair

Prices per pair. Carriage £2.50.

Dalesford System 1	£52.90
Dalesford System 2	£55.75
Dalesford System 3	£101.75
Dalesford System 4	£108.00
Dalesford System 5	£139.00
Dalesford System 6	£93.00
Eagle SK210	£13.90
Eagle SK215	£23.50
Eagle SK320	£33.50
Eagle SK325	£59.00
Eagle SK335	£79.90
Goodmans DIN20	£31.50
Goodmans Mezzo Twinkit	£51.95
Lowther PM6 Kit	£91.75
Lowther PM6 MKI Kit	£96.50
Peerless 1060	£66.95
Peerless 1070	£115.00
Peerless 1120	£129.50
Peerless 2050	£45.95
Peerless 2060	£60.95
Radford Studio 90	£154.00
Radford Monitor 270	£208.00
Radford Studio 270	£275.00
Radford Studio 360	£390.00
Richard Allan Twin	£29.90
Richard Allan Triple 8	£45.50
Richard Allan Triple 12	£55.90
Richard Allan Super Triple	£65.90
Richard Allan RA8	£42.75
Richard Allan RA82	£67.75
Richard Allan RA82L	£73.50
Seas Mini	£17.90
Seas 203	£35.50
Seas 302	£43.90
Seas 303	£73.90
Seas 503	£111.90
Wharfedale Denton 2XP	£26.95
Wharfedale Linton 3XP	£41.95
Wharfedale Glendale 3XP	£56.95

Everything in stock for the speaker constructor! BAF, long fibre wool, foam, crossovers, felt panels, components, etc. Large selection of grille fabrics (Send 15p in stamps for fabric samples) (Prices correct at 12/4/78)



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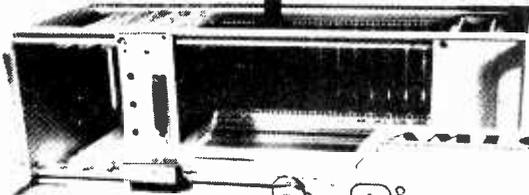
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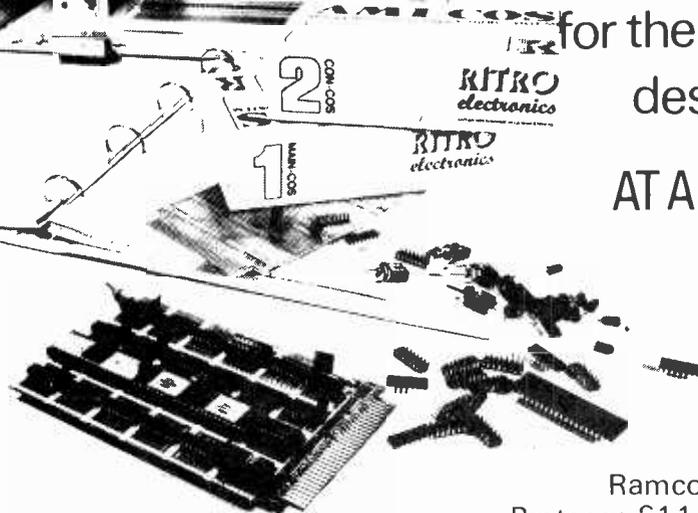
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Wilmslow, Cheshire.

AMI-COS microprocessor system



a modular 6800 system
for the amateur, student,
designer and lecturer

AT A REASONABLE PRICE!



Based on the Eurocard
format the kit prices are:
Maincos £82.50, Concos
£84, Manual £12.50,
Ramcos £60.70, Cosbus £29.00,
Protocos £11.50, Piacos £37.00, Sercos
£58.70 etc.

AMI-COS features modular flexibility. Each module is designed as a plug-in PCB, having identical serial contacts, and is connected to a common bus-line PCB, the COSBUS. The basic module MAINCOS incorporating the AMI S6800 CPU chip drives the binary command unit CONCOS. By using LED's and switches, the user will acquire a deeper knowledge and experience in the microprocessor field. Both modules are expandable with standard TV interfaces, cassette recorders, A/D converters, extra RAM (up to 65K x 8), and alpha-numeric or hexadecimal keyboards.

The manual supplied with MAINCOS and CONCOS contains the AMI 6800 hardware and the module documentation, with instructions.

All AMI-COS modules are available separately.

The application of AMI-COS is not limited to students or amateurs. Its flexibility makes it very useful for industrial applications. Co-operating with hardware and software consultants RITRO assures you of the right back up.

An AMI-COS brochure is available on request.

AMI-COS – YOUR LOGICAL FUTURE.

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WW — 121 FOR FURTHER DETAILS

QUALITY COMPONENTS BY RETURN



Linear Circuits by RCA, National etc.			TTL from National, Texas, ITT etc.			Transistors			Carbon Film Resistors							
709 DIL14	25p	LM322N	105p	7400	12p	7496	50p	AC125	18p	BC558	12p	ZTX304	25p	2N3704	8p	High stability, low noise. 0.25W 5% E12 series from 4.7 ohms to 10 Megohms. Any selection
741 DIL8	22p	LM1303	110p	7401	12p	7497	140p	AC126	18p	BC559	13p	ZTX310	13p	2N3705	8p	each 100+ 1000+ 5000
741 DIL14	35p	LM3900	50p	7402	12p	74100	90p	AC127	17p	BCY70	14p	ZTX311	14p	2N3706	8p	1p 0.9p 0.8p 0.75p
747C DIL14	50p	LM3909	60p	7403	12p	74104	50p	AC128	16p	BCY71	14p	ZTX314	22p	2N3707	8p	Special development pack of 10 of each value 4.7 ohms to 1 Megohm, a total of 650 resistors
748C DIL8	30p	MC1310P	150p	7404	13p	74105	40p	AC176	18p	BCY72	14p	ZTX341	21p	2N3708	8p	£5.70
CA3011	80p	MC1312P	150p	7405	13p	74107	25p	AD186	24p	BD115	52p	ZTX500	16p	2N3709	8p	Potentiometers
CA3014	130p	MC1314P	300p	7406	24p	74109	30p	AD191	36p	BD131	35p	ZTX501	16p	2N3710	8p	Carbon track. Log and Linear values.
CA3018	80p	MC1315P	520p	7407	24p	74110	35p	AD162	38p	BD132	35p	ZTX502	20p	2N3711	8p	5K - 2M2 single gang 26p
CA3020	160p	MC1330	100p	7408	14p	74111	45p	AF124	27p	BD133	44p	ZTX503	20p	2N3715	10p	5K - 2M2 single gang switched 8p
CA3028	125p	MC1458N	35p	7409	14p	74116	95p	AF125	27p	BD135	38p	ZTX504	20p	2N3719	10p	5K - 2M2 dual gang stereo 75p
CA3035	140p	MC1496N	60p	7410	12p	74118	82p	AF126	27p	BD136	36p	ZTX530	30p	2N3823	65p	Preset Potentiometers
CA3036	170p	NE555	25p	7411	15p	74119	140p	AF127	27p	BD137	39p	ZTX550	24p	2N3824	75p	Subminiature type available in horizontal or vertical mounting.
CA3042	170p	NE556	60p	7412	12p	74121	28p	AF139	36p	BD138	38p	ZTX696	32p	2N3866	55p	0.1W rating. 100 ohms to 2M 6p each
CA3043	180p	NE580	300p	7413	25p	74123	40p	AF239	40p	BD139	35p	ZTX697	12p	2N3903	8p	Special development of 5 of each value from 100 ohms to 2M, a total of 70 presets (please state vertical or horizontal) £3.95
CA3046	55p	NE561B	350p	7414	48p	74125	35p	BC107	8p	BD140	35p	ZTX698	28p	2N3904	8p	Ceramic Capacitors
CA3052	150p	NE562B	350p	7415	24p	74126	35p	BC107B	10p	BF244B	36p	ZTX699	50p	2N3905	8p	Miniature plate type. 50V PC mounting. Available from 22pF to 1000pF in E12 series and 1500pF to 0.047uF in E6 series
CA3054	115p	NE565A	120p	7416	24p	74132	50p	BC108	8p	BFX29	25p	ZTX706	13p	2N3906	30p	2p each
CA3075	180p	NE566V	150p	7417	24p	74151	50p	BC108B	8p	BFX84	23p	ZTX706A	13p	2N4037	8p	Mullard C280 series. 250V PC mounting.
CA3080	70p	NE587V	170p	7420	12p	74141	50p	BC108C	10p	BFX87	20p	ZTX708	20p	2N4058	12p	0.01, 0.015, 0.022, 0.033, 0.047, 0.068, 0.1, 5p; 0.15, 0.22, 7p; 0.33, 0.47, 10p; 0.68, 14p; 1.0, 17p; 2.2uF, 28p each.
CA3081	125p	SN76003N	200p	7421	22p	74142	200p	BC109	8p	BFX88	20p	ZTX914	22p	2N4059	12p	Tantalum Capacitors
CA3089	180p	SN76013N	140p	7422	18p	74145	58p	BC109C	10p	BFY50	15p	ZTX918	30p	2N4060	10p	Special development pack of 5 of each value £6.20
CA3090	400p	SN76023N	140p	7423	22p	74147	110p	BC147	7p	BFY51	15p	ZTX919	50p	2N4061	12p	0.1, 0.15, 0.22, 0.33, 0.47, 0.68, 1.0, 2.2 @ 35V 8p
CA3123	150p	SN76033N	200p	7424	24p	74150	70p	BC148	7p	BFY52	15p	ZTX920	50p	2N4062	10p	4.7 @ 25V, 6.8 and 10 @ 25V 13p
CA3130	90p	TA6621A	215p	7425	22p	74151	50p	BC149	8p	BU105	170p	ZTX929	25p	2N4063	32p	22 @ 16V, 47 @ 6V, 68 @ 3V, 100 @ 3V 16p
CA3140E	70p	TBA120S	65p	7426	24p	74153	50p	BC157	8p	BU208	160p	ZTX930	20p	2N4064	30p	Development pack 5 of each value £8.30
LM3001	130p	TBA540	200p	7427	24p	74153	50p	BC158	8p	MJ2955	98p	ZTX931	23p	2N4065	50p	Optoelectronics
LM301AN	28p	TBA641	240p	7428	22p	74155	52p	BC159	8p	MJ2955	98p	ZTX932	23p	2N4066	50p	LEDs
LM304H	70p	TBA800	70p	7433	32p	74156	52p	BC167	8p	MJF102	36p	ZTX933	38p	2N4067	10p	0.125 in Red 9p Green 15p Yellow 18p
LM308N	65p	TBA920	320p	7437	22p	74157	53p	BC168	8p	MPSA06	30p	ZTX934	54p	0A47 5p		
LM318N	125p	TCA2705Q	200p	7438	22p	74160	60p	BC169	8p	MPSA56	30p	ZTX935	54p	0A91 5p		
LM324N	50p	TDA1002	480p	7440	13p	74161	65p	BC169C	9p	TIP29	40p	ZTX936	54p	0A200 4p		
LM339	50p	TDA1022	570p	7441	52p	74162	65p	BC170	9p	TIP29A	44p	ZTX937	54p	1N914 4p		
LM3801	75p	TDA202D	320p	7442	43p	74163	65p	BC171	9p	TIP29B	40p	ZTX938	54p	1N916 4p		
LM381N	105p	ZN414	75p	7443	75p	74164	70p	BC172	7p	TIP29C	40p	ZTX939	54p	1N918 4p		
				7444	75p	74165	70p	BC173	9p	TIP30	80p	ZTX940	54p	1N4001 5p		
				7445	70p	74166	80p	BC177	14p	TIP30A	48p	ZTX941	54p	1N4002 4p		
				7446	55p	74167	180p	BC178	14p	TIP30B	55p	ZTX942	54p	1N4006 6p		
				7447	55p	74170	125p	BC179	14p	TIP30C	70p	ZTX943	54p	1N4148 3p		
				7448	58p	74172	400p	BC182	10p	TIP31	50p	ZTX944	54p			
				7450	14p	74173	90p	BC182L	10p	TIP31A	50p	ZTX945	54p			
				7451	14p	74174	68p	BC183	10p	TIP31B	50p	ZTX946	54p			
				7452	13p	74175	65p	BC183C	10p	TIP31C	65p	ZTX947	54p			
				7454	14p	74176	58p	BC184	10p	TIP32	55p	ZTX948	54p			
				7459	14p	74177	58p	BC184L	10p	TIP32A	60p	ZTX949	54p			
				7470	28p	74178	80p	BC207	10p	TIP32B	75p	ZTX950	54p			
				7472	24p	74181	145p	BC208	8p	TIP32C	80p	ZTX951	54p			
				7473	25p	74182	60p	BC209C	10p	TIP33	75p	ZTX952	54p			
				7474	25p	74185	110p	BC212	10p	TIP33A	80p	ZTX953	54p			
				7475	32p	74190	72p	BC212L	10p	TIP33B	103p	ZTX954	54p			
				7476	28p	74191	72p	BC213	10p	TIP33C	116p	ZTX955	54p			
				7483	80p	74192	64p	BC213L	10p	TIP34	95p	ZTX956	54p			
				7485	70p	74193	64p	BC214	10p	TIP34A	95p	ZTX957	54p			
				7486	24p	74194	80p	BC214L	10p	TIP34B	128p	ZTX958	54p			
				7489	145p	74195	55p	BC277	19p	ZTX107	14p	ZTX121	25p	95L05 70p		
				7490	32p	74196	55p	BC278	18p	ZTX108	14p	ZTX133	25p	95L12 70p		
				7491	45p	74197	55p	BC279	18p	ZTX109	14p	ZTX340	90p	95L15 70p		
				7492	35p	74198	110p	BC547	11p	ZTX300	16p	ZTX311	120p	LM309K 110p		
				7493	34p	74199	110p	BC548	10p	ZTX301	16p	ZTX342	135p	LM317K 300p		
				7494	80p			BC549	11p	ZTX302	23p	ZTX302	8p	LM323K 530p		
				7495	52p			BC550	14p	ZTX303	23p	ZTX303	8p	LM723 40p		

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New: Size TEK D12X

A wide range of small and miniature cases in A.B.S. suitable for anything from digital clocks to audio consoles. Low cost with discounts for quantity. **MODULOS** can be used for encapsulation and come with P.C. boards. **NUOVA** have a clear front panel. The very popular **DESKO** & **TEKO** are very widely used in both production and amateur use. **ALBA** are cases with grills top and bottom, for speakers or ventilation.

TEKO MODULOS

Lght mm	Wth mm	Hgt mm	Case	Panel	1 off	10	50
20	20	20	TEK L20 X	Pk 4	0.93	0.79	0.70
20	20	20	TEK L30 X	Pk 4	0.93	0.79	0.70
40	20	20	TEK L40 X	Pk 4	1.06	0.89	0.79
50	20	20	TEK L50 X	Pk 4	1.06	0.89	0.79
19	12.5	17.5	TEK S19 X	Pk 4	0.93	0.79	0.70
27	12.5	17.5	TEK S27 X	Pk 4	0.93	0.79	0.70
38	12.5	17.5	TEK S38 X	Pk 4	1.06	0.89	0.79
51	12.5	17.5	TEK S51 X	Pk 4	1.06	0.89	0.79
Asst of 8 pieces TEK S18 X Pk 8 1.87 1.58 1.40							

TEKO ALBA

Lght mm	Wth mm	Hgt mm	1 off	10	50
TEK A11	198	180	40		
TEK A12	198	180	55		
TEK A22	198	180	70		
TEK A23	198	180	90		
TEK A33	198	180	110		

DESKO

Dimensions	Panel	TEKO	1 off	10	50
80x50x30	TEK P1A	TEK P1P	0.88	0.58	0.51
105x65x40	TEK P2A	TEK P2P	1.01	0.86	0.76
155x90x50	TEK P3A	TEK P3P	1.49	1.27	1.12
210x125x70	TEK P4A	TEK P4P	2.48	2.11	1.86

MINIMUM ORDER £2.00

L	W	D	E	DESKO	1 off	10	50
161	95	60	40	TEK 362	1.65	1.40	1.24
215	130	75	45	TEK 363	2.48	2.11	1.86
311	169	90	50	TEK 364	5.21	4.43	3.91

TRADE ENQUIRIES INVITED

All West Hyde cases are available with substantial discounts for quantities. Most cases have discounts at 5, 10 and 25 off with discounts up to 33% at 100 off. Prices include P & P and are less 10% if collected on first two price breaks on cases only. Send for catalogue. Prices correct at press date.

WEST HYDE DEVELOPMENTS LIMITED, Unit 9, Park Street Industrial Estate, AYLESBURY, BUCKS. HP20 1ET. Phone: Aylesbury (0296) 20341. Telex: 83570

WW-078 FOR FURTHER DETAILS



EASY BUILD SPEAKER DIY KITS

Specially designed by RT-VC for cost-conscious hi-fi enthusiasts, these kits incorporate two teak-simulate enclosures, two EMI 13" x 8" (approx.) woofers, two tweeters and a pair of matching crossovers. Supplied complete with an easy-to-follow circuit diagram, and crossover components.

£2800
STEREO PAIR Input 15 watts rms, 30 watts peak, each unit. Cabinet size 20" x 11" x 9 1/2" (approx.).
 + p & p £5.50
SPEAKERS AVAILABLE WITHOUT CABINETS.
 It's the units which we supply with the enclosures illustrated. Size 13" x 8" (approx.) woofer (EMI). £1700 per tweeter, and matching crossover components. stereo pair Power handling 15 watts rms, 30 watts peak. + p & p £3.40

BUILT AND READY TO PLAY

SPEAKERS Two models—Duo IIb, teak veneer, 12 watts rms, 24 watts peak, 18 1/2" x 13 1/2" x 7 1/2" (approx.). Duo III, 20 watts rms, 40 watts peak, 27" x 13" x 11 1/2" approx.
Duo III £17 PER PAIR p & p £6.50
Duo IIb £52 PER PAIR p & p £7.50

EASY TO BUILD



WITH SPEAKERS NOT TO SCALE



RECORD PLAYER KIT

for the D.I.Y. man who requires a stereo unit at a budget price, comprising ready assembled stereo amp module, Garrard auto/manual deck with cueing device, pre-cut and finished cabinet work. Output 4 watts per channel, phones socket and record/replay socket including 2 SPHERICAL HI-FI speakers. **£19.95** p & p £4.05

AM/FM STEREO TUNER AMPLIFIER CHASSIS COMPLETE

Ready built. Designed in a slim form for compact, modern installation.
Rotary Controls Vol/Dn/DH Bass, Treble, Balance.
Push Buttons for Gram, Tape VHF, MW, LW and 5 button rotary selection switch.
Power Supply Selenium Bridge—35V DC from 210-250V AC, 50Hz input.
Aerial Ferrite 8" x 1/2" built into chassis for LW and MW plus flying lead for FM aerial.
Power Output 5 watts per channel. Sine at 2% THD into 15 Ohm 7 watts speech and music.
Tape Sensitivity Playback 400mV/30K OHM for max output. Record 200mV/50K output available from 25KHz (150mV/100K) deviation FM signal. **Frequency Range (Audio)** 50Hz to 17 KHz within ±10dB.
Radio FM sensitivity for 3dB below limiting better than 10 uV AM sensitivity for 20 dB S/N. MW 350 uV/Metre. LW 1 mV/Metre. Size approx. length 16" x height 2 1/2" x depth 4 1/2". **£19.95** P & P £2.50

VALUE FOR PERSONAL SHOPPERS

- 160 16 VOLT MAINS TRANSFORMER, 2 1/2 amp. **£2.50**
- BSR Record auto deck on plinth with stereo cartridge, ready wired **£11.95**
- LED 5 function men's digital watch stainless steel finish **£5.95**
- LED 5 function men's digital watch stainless steel finish **£7.95**
- LCD 8 function CHRONOGRAPH men's digital watch stainless steel finish **£12.95**
- STEREO CASSETTE record/replay fully built P.C. board. Used, without guarantee. (Ex Equipment) **£1.95**
- 125 Watt Power Amp Module **£13.95**
- Mains power supply parts **£3.50**
- 100K Multiturn Varicap tuning pots, 6 for **£1.00**
- MUSIC CENTRE CABINET with hinged smoke acrylic top, finished in natural teak veneers, size 30 1/2" x 14 1/2" x 7 1/2" approx. **£5.95**
- MULLARD Built power supply **£1.50**
- DECCA DC 1000 Stereo Cassette P.C.B. complete with switch oscillator coils and tape heads **£2.95**
- IMF TLS 80 Monitor loudspeaker cabinet size approx. 43 1/2" x 15 1/2" x 15 1/2" **£24.95**
- DECCA 20w Stereo speaker kit comprising 2 8" approx. bass units + 2 3 1/2" approx. tweeter inc crossovers **£20.00**
- 2 BAND CLOCK RADIO, Mains operated with sleep control **£10.95**
- VIDEOMASTER Super Score TV Game with pistol mains operation **£14.95**
- VIDEOMASTER Door Tunes (24 different titles) **£12.95**
- Micro cassette tape recorder **£13.95**
- 7" TAPE TRANSPORT Mechanism—a selection of models from **£8.95**

Mullard

AUDIO MODULES IN BARGAIN PACKS CURRENT CATALOGUE

PRICE £ AT OVER 25 PER PACK

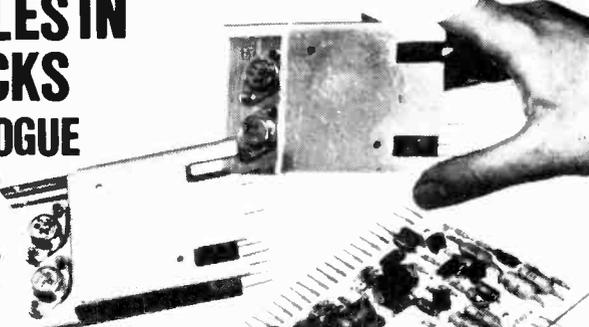
SEE OUR PRICES

- 1** PACK 1. 2 x LP1173 10w. RMS output power audio amp modules, + 1 LP1182/2 Stereo pre amp for ceramic and auxiliary input. **OUR PRICE p+p £1.00 £4.95**
- 2** PACK 2. 2 x LP1173 10w. RMS output power audio amp modules + 1 LP1184/2 Stereo pre amp for magnetic, ceramic and auxiliary inputs **illu. OUR PRICE p+p £1.00 £7.45**
- 3** PACK 3. 1 x LP1179/2 FM Tuning head with AM gang, 1 x LP1165/1 AM/FM IF module, 2 x LP1173/10w. RMS output power audio amp modules + 1 LP1182/2 Stereo pre amp for ceramic and auxiliary input. **OUR PRICE p+p £1.00 £9.95**

TRADE ENQUIRIES INVITED

THIS MONTH'S OFFER

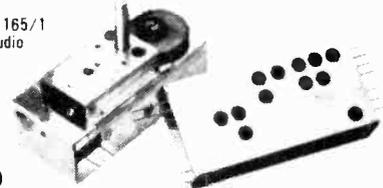
added to our bargain packs. When you buy Pack 3 at £9.95, together with a mains transformer at £1.95 and a set of controls for 95p, you receive FREE a Mullard LP1400 Decoder to match. Listed at **£11.90** P&P £2.51 **£12.85**



ACCESSORIES

Suitable power supply parts including mains transformer, rectifier, smoothing and output capacitors. **£1.00 p+p £1.95**

Recommended set of rotary stereo controls comprising BASS, TREBLE, VOLUME and BALANCE. **p+p 50p 95p**



20 x 20 WATT STEREO AMPLIFIER **£29.90**
 Superb Viscount IV unit in teak-finished cabinet. Silver fascia with aluminium rotary controls and pushbuttons, red mains indicator and stereo jack socket. Function switch for mic, magnetic and crystal pick-ups, tape, tuner, and auxiliary. Rear panel features two mains outlets, DIN speaker and input sockets, plus fuse. **20 + 20 watts rms, 40 + 40 watts peak.**
30 x 30 WATT AMPLIFIER KIT
 For the experienced constructor complete in every detail. Similar facilities as Viscount IV amplifier. **60 + 60 peak £29.00** p & p £2.50
AVAILABLE NOW built and fully tested with output **£39.00** 30 + 30 watts rms, 60 + 60 peak. p & p £2.50
SPECIAL OFFER
PACKAGE PRICE WITH 30 x 30 KIT Mk. II version, operates into 4 to 15 OHMS speakers. Designed by R & TVC for the experienced constructor. Complete in every detail, facilities as Viscount IV amplifier. **60 + 60 peak.** Supplied with 2 Goodmans Compact 12" bass woofer with cropped 14,000 Gauss Magnet. 30 watt rms, handling +3/4" approx. tweeters and crossovers. **£49.00** p & p £4.00



50 WATT MONO DISCO AMP **£29.95** P&P £2.50
 Size approx. 13 1/4" x 5 1/4" x 6 1/4"
 50 watts rms, 100 watts peak output. Big features include two disc inputs, both for ceramic cartridges, tape input and microphone input. Level mixing controls fitted with integral push-pull switches. Independent bass and treble controls and master volume.
SPECIAL OFFER. The above 50 watt amp plus 4 Goodmans Type 8P, 8" speakers. Package price **£45.00 + £4.00 P&P.**



70 & 100 WATT MONO DISCO AMP **£29.95**
 Size approx. 14" x 4" x 10 1/2"
 Brushed aluminium fascia and rotary controls.
 Five vertical slide controls—master volume, tape level, mic level, deck level, PLUS INTER-DECK FADER for perfect graduated change from record deck No. 1 to No. 2, or vice versa. Pre-fade level control 70 watt (PFL) lets YOU hear next disc before fading. 100 watt output level. **100 watt £65**
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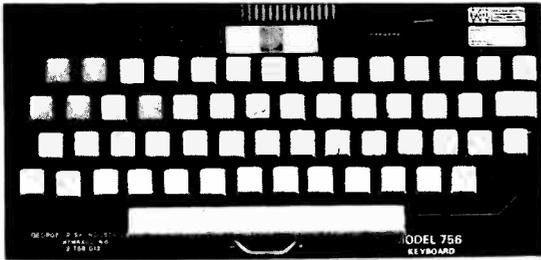
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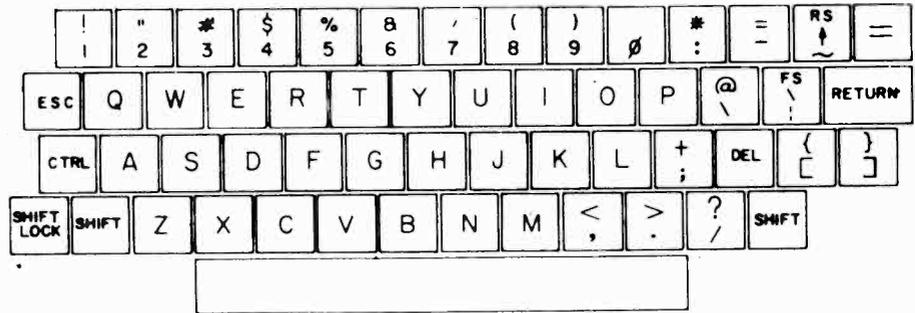
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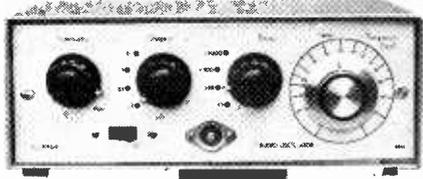
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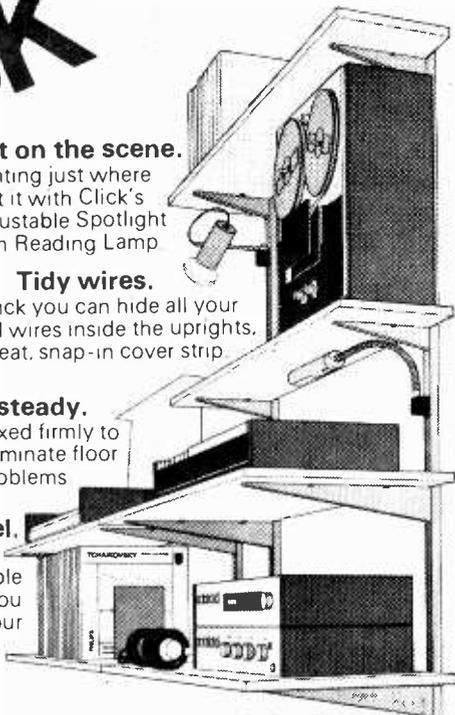
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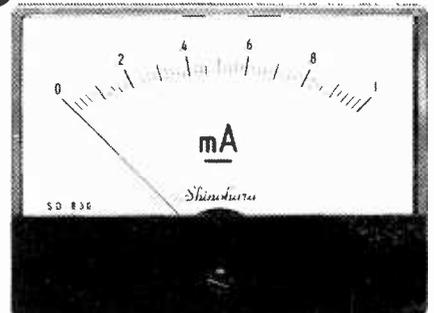
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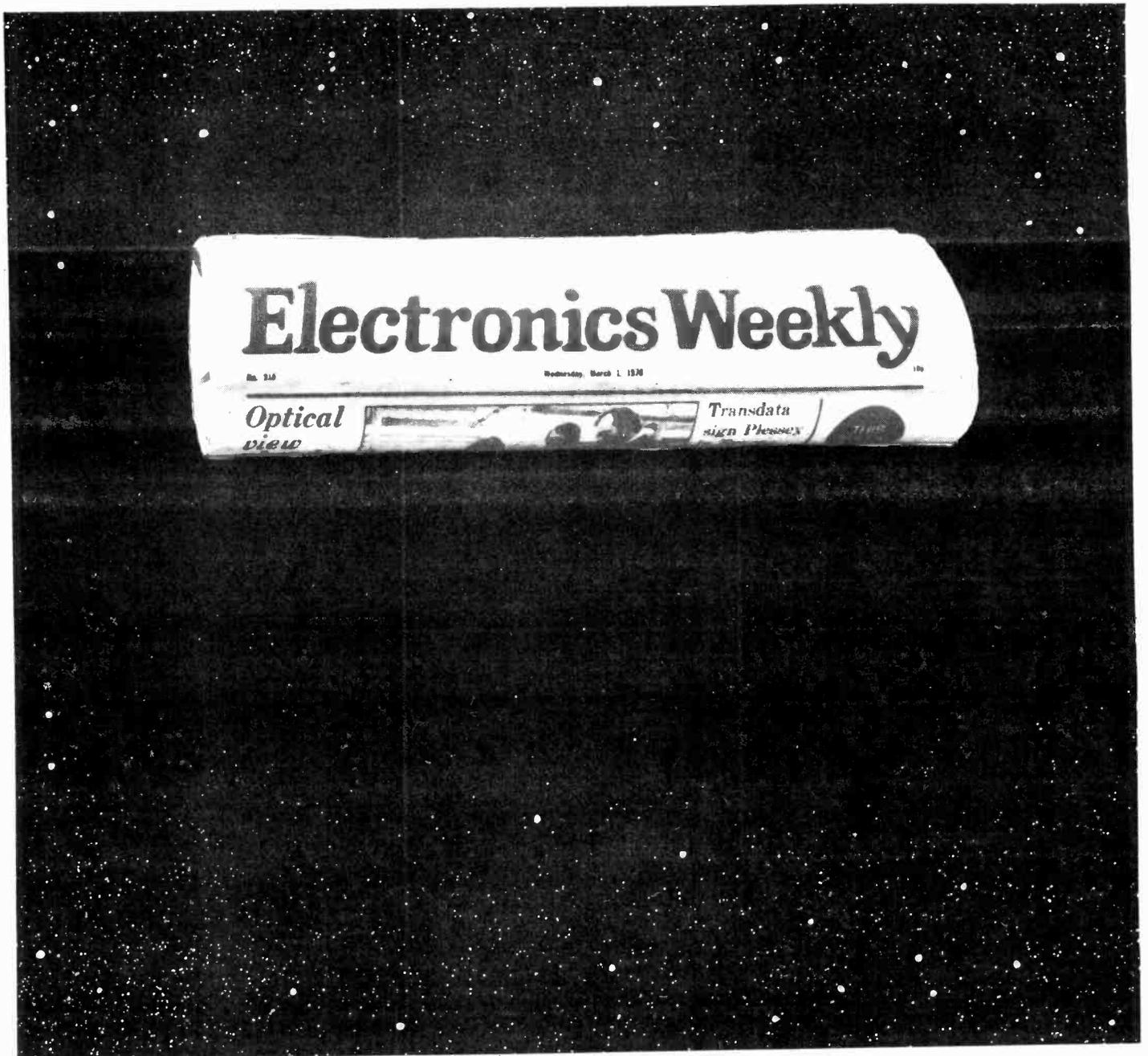


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8C107	6p	8C251	10p	OC45	12p	ZTX500	8p	2N2906	12p
8C108	6p	8CY70	12p	OC71	9p	ZTX501	10p	2N2906A	14p
8C109	6p	8CY71	12p	OC72	12p	ZTX502	12p	2N2907	12p
8C118	10p	8CY72	12p	OC75	10p	2N696	10p	2N2907A	13p
8C147	8p	8D115	40p	OC81	14p	2N697	10p	2N2926G	8p
8C148	8p	8D131	35p	TIP29A	35p	2N706	7p	2N2926Y	7p
8C149	8p	8D132	37p	TIP29B	36p	2N706A	8p	2N3053	12p
8C154	16p	8F115	17p	TIP29C	38p	2N708	8p	2N3055	35p
8C157	9p	8F167	19p	TIP30A	36p	2N1302	12p	2N3702	7p
8C158	9p	8F173	20p	TIP30B	37p	2N1303	15p	2N3703	7p
8C159	9p	8F180	15p	TIP30C	38p	2N1304	15p	2N3704	6p
8C169C	10p	8F181	15p	TIP31A	32p	2N1307	18p	2N3903	11p
8C170	6p	8F182	25p	TIP31B	33p	2N1308	22p	2N3904	11p
8C171	6p	8F183	25p	TIP31C	34p	2N1309	22p	2N3905	11p
8C172	6p	8F184	25p	TIP32A	34p	2N1613	15p	2N3906	11p
8C173	7p	8F185	25p	TIP32B	35p				

POTENTIOMETERS

Slider 40mm TRAVEL	
Order No	
16191 6 x 470 Ohm	LIN Single 40p
S24 6 x 1K	LIN Single 40p
S25 6 x 5K	LIN Single 40p
16192 6 x 10K	LIN Single 40p
16193 6 x 22K	LIN Single 40p
16195 6 x 47K	LOG Single 40p
16194 6 x 47K	LIN Single 40p
S27 6 x 100K	LIN Single 40p
S28 6 x 100K	LOG Single 40p
S29 6 x 500K	LOG Single 40p

Slider 60mm TRAVEL

S30 6 x 2.5K	LOG Single 40p
S32 6 x 50K	LOG Single 40p
S33 6 x 250K	LOG Single 40p
S34 4 x 5K	LOG Dual 40p
S36 4 x 100K	LOG Dual 40p
S37 4 x 1.3 MEG	LOG Dual 40p
S94 6 x 220K LIN Single	40p
S95 6 x 100K LOG Single	40p
S96 6 x 500K LIN Single	40p

S38 MIXED SLIDER POTS - VARIOUS VALUES AND SIZES - OUR MIX ONLY £1.00

S39 6 x CHROME SLIDER KNOBS 40p

WIREWOUND

S90 Wirewound Pots Linear 1Watt rating. Mixed useful values 5 for £1.00

COMPONENT PAKS

SPECIAL OFFER!

Order No	Quantity	
16168	5 pieces Assorted Ferrite rods	40p
16169	2 pieces Tuning gangs MW/LW	40p
16170	50 metres Single strand wire assorted wire	40p
16171	10 Reed switches	40p
16172	3 Micro switches	40p
16176	20 Assorted electrolytics Trans types	40p
16177	1 pack Assorted hardware nuts/bolts, etc	40p
16179	20 Assorted tag strips and panels	40p
16180	15 Assorted control knobs	40p
16184	15 Assorted Fuses 100mA-5 amp	40p
16188	60 1/2W resistors mixed values	40p
16187	30 metres stranded wire assorted colours	40p

S100	120 1/2 watt resistors Pre-formed 1978 Prod	Our mix 60p
S101	120 1/2 watt resistors Pre-formed 1978 Prod	Mixed values 60p
S102	250 1/2 watt resistors Range 100ohms-1 meg	£2.00
S103	220 1/2 watt resistors Range 100ohms-10meg	£2.00
S104	60 Low ohms 1/2 watt res 10-100ohms	60p
S105	40 Low ohm 1/2 watt resistors. 22-82 ohms	60p
S106	25 Mixed wirewound resistors	60p
S107	20 Tantulum bead caps. 22-100mF	Our mix £1.00
S108	High quality electrolytics 10mF-500mF. voltage range 15-50v	Our mix 40 for £1.00
16204	C280 Pak Contains 50 metal foil caps	£1.00
3136	Ribbon cable flat standard 15-way multi-coloured PVC insulated, stranded tin copper. 1m	25p

SILICON POWER TRANS. NPN

S97	8D371 2amp 1 2w 60Vceo Hfe 40 400. Case T092 with heat tab 5 for	60p
S98	2N5293 RCA 36w 4amps 75Vceo Hfe 30-120	5 for £1.00

SILICON BRIDGE RECTS.

S99	Mixed pak 2-5amp 50-600v all coded	4 for £1.00
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PC BOARD

S110	Mixed Bundle PCB. Fibreglass/paper single & double sided	Fantasy value 75p
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I.C. SOCKET PAKS

No S61	11 x 8-pin DIL Sockets	£1.00
No S62	10 x 14-pin DIL Sockets	£1.00
No S63	9 x 16-pin DIL Sockets	£1.00
No S64	4 x 24-pin DIL Sockets	£1.00
No S70	1 x 28-pin DIL Sockets	£1.00

MAMMOTH I.C. PAK

Approx 200 Pieces Assorted fall-out integrated circuits, including Logic 74 series Linear Audio and D.T.L. Many coded devices but some unmarked - you to identify
Order No 16223 £1.00

ZENER PAKS

No S55	20 mixed values 400mW Zener diodes 3-10V	£1.00
No S56	20 mixed values 400mW Zener diodes 11-33V	£1.00
No S57	10 mixed values 1W Zener diodes 3-10V	£1.00
No S58	10 mixed values 1W Zener diodes 11-33V	£1.00

DIODES

Type	Price	Type	Price	Type	Price	Type	Price	Type	Price
AA119	£0.05	8AX16/		8Y216	£0.30	OA85	£0.07	IS44	£0.03
AA213	£0.04	OA202	£0.05	8Y217	£0.28	OA90	£0.06		
8A100	£0.06			8Y218	£0.28	OA91	£0.07	1N5400	£0.10
8A115	£0.05	8Y100	£0.15	8Y219	£0.28	OA95	£0.07	1N5401	£0.11
8A144	£0.05	8Y127	£0.10					1N5402	£0.12
8A148	£0.10	8Y210	£0.32	OA47	£0.05	IN34	£0.05	1N5404	£0.13
8A173	£0.10	8Y211	£0.32	OA70	£0.05	IN60A	£0.06	1N5406	£0.16
8AX13/		8Y212	£0.32	OA79	£0.07	IN914	£0.04	1N5407	£0.17
OA200	£0.05	8Y213	£0.30	OAB1	£0.07	IN414B	£0.04	1N5408	£0.19

LINEAR I.C.s

TB800 12 pin QIL	£0.75	UA711C T099	£0.25	UA748 T099	£0.28
TB810 12 pin QIL	£1.00	UA703 T099 (Plastic)	£0.20	72558 (Dual 74B) T099	£0.45
TB820 14 pin QIL	£0.80	741P 8 pin DIL	£0.18	MC1310P 14 pin DIL	£1.25
LM380 14 pin QIL	£0.80	72741 14 pin DIL	£0.20	76115 14 pin QIL	£1.25
LM381 14 pin DIL	£1.35	UA741C T099	£0.20	NE555 8 pin DIL	£0.25
72709 14 pin DIL	£0.28	72747 14 pin DIL	£0.55	NE556 14 pin DIL	£0.60
UA709 T099	£0.28	748P 8 pin DIL	£0.28	SL414A 10 pin	£1.80

NEW CONSIGNMENT ZN 414 RADIO CHIP 75p

OPTOELECTRONICS

Displays	No 1510 707 LED Display	70p each	2nd QUALITY LED PAKS	No 1507 10 x LEDs Assorted	75p
	No 1511 747 LED Display	£1.50 each	LED CLIPS	No 1508/1 125	5 for 12p
				No 1508/2 2	5 for 15p

LEDs	No S51 Red TIL209 (5 x 125)	50p
	No S52 Red FLV117 (5 x 2)	50p
	No 1502 Green 125	18p each
	No 1505 Green 2	18p each
	No 1503 Yellow 125	18p each
	No 1506 Yellow 2	18p each
	No S82 Clear 2 (illuminating red)	12p

SPECIAL REDUCTIONS

No 1514 NORP 12	45p each
No S76 OCP71	5 for £1.00
No S83 5 NIXIE Tubes ITT 5B70 ST	£2.00

(including Data)

SILICON SOLAR CELLS

S109 Output 5mA at 1/2 volt 50p each

P.O. RELAYS

S85 - 2 off Post Office relays 40p

CMOS ICs

Type	Price	Type	Price	Type	Price
CD4000	£0.14	CD4022	£0.80	CD4046	£0.95
CD4001	£0.16	CD4023	£0.18	CD4047	£0.75
CD4002	£0.16	CD4024	£0.64	CD4049	£0.46
CD4006	£0.80	CD4025	£0.18	CD4050	£0.46
CD4007	£0.17	CD4026	£1.85	CD4054	£0.95
CD4008	£0.80	CD4027	£0.48	CD4055	£1.60
CD4009	£0.50	CD4028	£0.80	CD4056	£1.15
CD4011	£0.50	CD4029	£0.95	CD4069	£0.32
CD4011	£0.18	CD4030	£0.46	CD4070	£0.32
CD4012	£0.17	CD4031	£1.80	CD4071	£0.20
CD4013	£0.42	CD4035	£1.40	CD4072	£0.20
CD4015	£0.80	CD4037	£0.78	CD4081	£0.20
CD4016	£0.42	CD4040	£0.78	CD4082	£0.20
CD4017	£0.80	CD4041	£0.68	CD4510	£1.10
CD4018	£0.85	CD4042	£0.68	CD4511	£1.25
CD4019	£0.45	CD4043	£0.78	CD4516	£1.10
CD4020	£0.95	CD4044	£0.78	CD4518	£1.10
CD4021	£0.85	CD4045	£1.15	CD4520	£1.10

AUDIO LEADS

Order No		
117	AC Mains connecting lead for cassette recorders and radios	45p
118	5 pin DIN headphone plug to stereo socket	78p
119	2 x 2 pin plug to inline stereo socket for headphones	60p
120	20ft of coiled guitar lead	£1.15
124	3 pin to 3 pin DIN plug	50p
125	Audio lead 5 pin plug to 5 pin DIN plug	50p
126	Audio lead 5 pin DIN plug to tinned open ends	50p
127	Audio lead 5 pin DIN plug to 4 phono plugs	90p
129	Audio lead 5 pin plug to 5 pin DIN plug Mirror image	70p
130	5 metre lead 2 pin DIN plug to pin DIN inline socket	45p
132	10 metre lead 2 pin DIN plug	65p

FET

2N3819	15p
2N5458	18p

AUDIO PLUG AND SOCKET PAKS

Order No.		
S1	5 x 3 5mm Plastic Jack Plugs	40p
S2	5 x 2 5mm Plastic Jack Plugs	40p
S3	4 x Std Plastic Jack Plugs	50p
S4	2 x Stereo Jack Plugs	30p
S5	5 x 5 Pin 180 DIN Plugs	50p
S6	8 x 2 Pin Loudspeaker Plugs	50p
S7	6 x Phono Plugs Plastic	50p
S8	5 x 3 5mm Chassis Sockets (switched)	25p
S9		



NEW STYLING

NRDC-AMBISONIC UHJ



SURROUND SOUND DECODER

The **first ever** kit specially produced by Integrex for this British NRDC backed surround sound system which is the result of 7 years' research by the Ambisonic team. W.W. July, Aug., '77.

The unit is designed to decode not only UHJ but virtually all other 'quadrophonic' systems (Not CD4), including the new BBC HJ 10 input selections.

The decoder is linear throughout and does not rely on listener fatiguing logic enhancement techniques. Both 2 or 3 input signals and 4 or 6 output signals are provided in this most versatile unit. Complete with mains power supply, wooden cabinet, panel, knobs, etc.

Complete kit, including licence fee **£45.00 + VAT**
or ready built and tested **£61.50 + VAT**



INTRUDER 1 RADAR ALARM

With Home Office Type approval

As in "Wireless World", designed by Mike Hosking. 240V ac mains operated and disguised as a hardbacked book. Detection range up to 30 feet.

Complete exclusive designer approved kit **£46.00 + VAT**
or ready built and tested, **£54.00 + VAT**

Wireless World DolbyTM noise reducer

Trademark of Dolby Laboratories Inc.



Featuring:

- switching for both encoding (low-level h.f. compression) and decoding
- a switchable f.m. stereo multiplex and bias filter.
- provision for decoding Dolby f.m. radio transmissions (as in USA).
- no equipment needed for alignment.
- suitability for both open-reel and cassette tape machines.
- check tape switch for encoded monitoring in three-head machines.

Typical performance

Noise reduction better than 9dB weighted.
Clipping level 16.5dB above Dolby level (measured at 1% third harmonic content)

Harmonic distortion 0.1% at Dolby level typically 0.05% over most of band, rising to a maximum of 0.12%

Signal-to-noise ratio 75dB (20Hz to 20kHz, signal at Dolby level) at Monitor output

Dynamic Range >90dB

30mV sensitivity.

Complete Kit **PRICE: £39.90 + VAT**

Also available ready built and tested **Price £54.00 + VAT**

Calibration tapes are available for open-reel use and for cassette (specify which) **Price £2.20 + VAT**

Single channel plug-in DolbyTM PROCESSOR BOARDS (92 x 87mm) with gold plated contacts are available with all components **Price £8.20 + VAT**

Single channel board with selected fet **Price £2.50 + VAT**

Gold Plated edge connector **Price £1.50 + VAT**

Selected FETs **60p** each + VAT, **100p** + VAT for two, **£1.90** + VAT for four

Please add VAT @ 12½% unless marked thus, when 8% applies (or current rates)

We guarantee full after-sales technical and servicing facilities on all our kits, have you checked that these services are available from other suppliers?



INTEGREX LTD.

Please send SAE for complete lists and specifications
**Portwood Industrial Estate, Church Gresley,
Burton-on-Trent, Staffs DE11 9PT
Burton-on-Trent (0283) 215432 Telex 377106**

INTEGREX

S-2020TA STEREO TUNER/AMPLIFIER KIT

SOLID MAHOGANY CABINET

A high-quality push-button FM Varicap Stereo Tuner combined with a 24W r.m.s. per channel Stereo Amplifier.



Brief Spec. Amplifier Low field Toroidal transformer, Mag. input, Tape In/Out facility (for noise reduction unit, etc.), THD less than 0.1% at 20W into 8 ohms. Power on/off FET transient protection. All sockets, fuses, etc., are PC mounted for ease of assembly. Tuner section uses 3302 FET module requiring no RF alignment, ceramic IF, INTERSTATION MUTE, and phase-locked IC stereo decoder. LED tuning and stereo indicators. Tuning range 88–104MHz. 30dB mono S/N @ 1.2 μ V. THD 0.3%. Pre-decoder 'birdy' filter. **PRICE: £58.95 + VAT**

NELSON-JONES MK. I STEREO FM TUNER KIT

A very high performance tuner with dual gate MOSFET RF and Mixer front end, triple gang varicap tuning, and dual ceramic filter/dual IC IF amp.



Brief Spec. Tuning range 88–104MHz. 20dB mono quieting @ 0.75 μ V. Image rejection – 70dB. IF rejection – 85dB. THD typically 0.4%. IC stabilized PSU and LED tuning indicators. Push-button tuning and AFC unit. Choice of either mono or stereo with a choice of stereo decoders.

Compare this spec. with tuners costing twice the price.

Mono £32.40 + VAT
With ICPL Decoder £36.67 + VAT
With Portus-Haywood Decoder £39.20 + VAT

Please send for details of the Nelson-Jones Mk. II Kit as in this month's W.W.



Sens. 30dB S/N mono @ 1.2 μ V
 THD typically 0.3%
 Tuning range 88–104MHz
 LED sig. strength and stereo indicator

STEREO MODULE TUNER KIT

A low-cost Stereo Tuner based on the 3302 FET RF module requiring no alignment. The IF comprises a ceramic filter and high-performance IC Variable INTERSTATION MUTE: PLL stereo decoder IC. Pre-decoder 'birdy' filter Push-button tuning

PRICE: Stereo £31.95 + VAT



S-2020A AMPLIFIER KIT

Developed in our laboratories from the highly successful "TEXAN" design. PC mounting potentiometers, switches, sockets and fuses are used for ease of assembly and to minimize wiring Power 'on/off' FET transient protection.

Typ Spec. 24+24W r.m.s. into 8-ohm load at less than 0.1% THD. Mag. PU input S/N 60dB. Radio input S/N 72dB. Headphone output. Tape In/Out facility (for noise reduction unit, etc.). Toroidal mains transformer.

PRICE: £33.95 + VAT

ALL THE ABOVE KITS ARE SUPPLIED COMPLETE WITH ALL METALWORK, SOCKETS, FUSES, NUTS AND BOLTS, KNOBS, FRONT PANELS, SOLID MAHOGANY CABINETS AND COMPREHENSIVE INSTRUCTIONS

BASIC NELSON-JONES TUNER KIT	£14.28 + VAT	PHASE-LOCKED IC DECODER KIT	£4.47 + VAT
BASIC MODULE TUNER KIT (stereo)	£16.75 + VAT	PUSH-BUTTON UNIT	£5.00 + VAT
PORTUS-HAYWOOD PHASE-LOCKED STEREO DECODER KIT			£8.00 + VAT

Z & I AERO SERVICES LTD.

Head Office: 44A WESTBOURNE GROVE, LONDON W2 5SF
Tel.: 727 5641 Telex: 261306

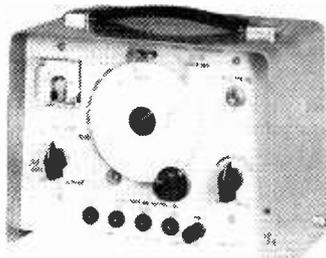
RETAIL SHOP
85 TOTTENHAM COURT ROAD, W.1
Tel 580-8403

R.C. OSCILLATOR G3-36A

Made in USSR

Portable transistorized R-C oscillator providing sinewave and 50/50 squarewave. Four separate output sockets give attenuation ratio of 1, 10, 100 and 1000. Output 0-5 volts R M S. Frequency range 20Hz-200KHz in four bands Output impedance 600Ω for sinewave and 4000Ω for squarewave. Harmonic distortion 1-2% Power supplies 200-240V AC

Price **£37.00**
Packing and delivery **£2.00**
(VAT 8% to be added to the above figure)



TAUT SUSPENSION MULTIMETERS

Made in USSR



TYPE	U4313	U4315
Sensitivity D C	20,000 Ω p v	20,000 Ω p v
Sensitivity A C	2,000 Ω p v	2,000 Ω p v
D C Current	60 μ A-1.5A	50 μ A-2.5A
A C Current	0.6mA-1.5A	0.5mA-2.5A
D C Volts	75mV-600V	75mV-1000V
A C Volts	15V-600V	1V-1000V
Resistance	1K-1M	300Ω-500kΩ
Capacity	0.5 μ F	0.5 μ F
Accuracy	1.5% D C 2.5% A C	2.5% D C 4% A C

Price complete with pressed steel carrying case and test leads. **£17.50** **£15.85**
Packing and postage **£1.50** **£1.50**
Plus VAT at 8%

FULLY GUARANTEED



VALVES

0A2	0.55	6J5GT	0.80	ECC88	0.75	EF41	0.75	EY51	0.60	PL508	1.30
0A3	0.75	6J6	0.55	ECC89	0.80	EF80	0.40	EY81	0.50	PL509	1.75
0B2	0.60	6L6GT	0.85	ECC189	0.80	EF85	0.48	EY87	0.50	PL802	2.80
0C3	0.75	6SL7GT	0.70	ECF80	0.60	EF86	0.60	EY88	0.55	PY31	0.50
1B3GT	0.65	6SN7GT	0.70	ECF82	0.55	EF97	0.70	EY500A	1.50	PY33	0.63
1R5	0.50	12AL5	0.65	ECF86	0.80	EF98	0.90	EZ80	0.50	PY80	0.60
1X2B	1.20	12AQ5	0.55	ECF200	0.90	EF183	0.70	EZ81	0.50	PY81	0.70
5R4GY	1.10	12AT6	0.60	ECF201	0.90	EF184	0.70	KT66	4.50	PY82	0.55
5U4G	0.60	12AT7	0.50	ECF801	0.95	EFL200	1.20	KT88	5.80	PY83	0.70
5U4GB	0.95	12AU6	0.65	ECF82	0.95	EL36	0.95	PC88	0.85	PY88	0.75
5V4G	0.60	12AU7	0.47	ECH42	1.10	EL41	0.80	PC88	0.85	PY500A	1.30
5Y3GT	0.65	12AV6	0.85	ECH81	0.55	EL81	0.65	PC92	0.85	TT21	7.80
5Z4GT	0.65	12AV7	1.00	ECH82	0.60	EL82	0.60	PC96	0.50	TT22	7.80
6B4	0.55	12AX7	0.55	ECH83	0.60	EL83	0.60	PC97	0.95	UABC80	0.58
6AJS	0.65	12AY7	0.85	ECH84	0.55	EL83	0.60	PC900	1.00	UAF41	0.80
6AK5	0.55	12BA6	0.65	ECH200	0.80	EL84	0.45	PC84	0.50	UAF42	0.70
6AL5	0.40	12BE6	0.80	ECL80	0.60	EL86	0.75	PC85	0.60	UBC41	0.70
6AM6	0.70	12BH7	0.75	ECL81	0.75	EL95	0.70	PC89	0.65	UBC81	0.60
6AQ5A	0.85	12X4	0.50	ECL82	0.60	EL504	0.60	PC189	1.00	UBF80	0.60
6AS5	0.75	19AQ5	0.75	ECL83	1.15	EM80	0.85	PCF80	0.65	UBF89	0.60
6AS6	1.00	35A3	0.70	ECL84	0.70	EM81	0.60	PCF82	0.45	UCC84	0.75
6AT6	0.75	35B5	0.65	ECL85	0.65	EM83	0.50	PCF84	0.65	UCC85	0.55
6AV6	0.75	35C5	0.70	ECL86	0.85	EM84	0.60	PCF86	0.75	UCC80	0.75
6AW8A	0.75	35W4	0.70					PCF88	0.75	UCH81	0.65
6AU6	0.50	50C5	1.00					PCF201	1.10	UCL81	0.70
6AV6	0.75	EABC80	0.55					PCL81	0.65	UCL82	0.75
6BA6	0.45	EAF42	0.70					PCL82	0.80	UCL83	0.80
6BE6	0.48	EAF801	0.70					PCL84	0.75	UCL83	0.80
6BJ6	1.20	EBC41	0.75					PCL86	0.85	UF41	1.00
6BN6	0.80	EBC81	0.70					PCL805	0.75	UF80	0.50
6BZ6	0.65	EBF80	0.50					P0510	3.35	UF85	0.50
6BZ7	0.70	EBF83	0.45					PL36	1.10	UF89	0.55
6C4	0.55	EBF89	0.50					PL38	0.65	UL41	0.80
6C86	0.55	EC86	0.75					PL81	0.80	UL84	0.85
6C06	2.20	EC88	0.75					PL82	0.55	UM80	0.60
6GK5	0.70	EC91	2.80					PL83	0.50	UM84	0.45
6GK6	0.90	ECC84	0.60					PL84	0.75	UY42	0.60
6J4	1.20	ECC85	0.48					PL95	0.70	UY82	0.60
								PL504	1.05	UY85	0.60

All prices are exclusive of VAT (12½%)

When ordering by post please add (unless otherwise indicated) 30p in £ for packing and postage, plus appropriate rate of VAT

Minimum order charge for approved credit customers is £20.00. Any order below £20.00 (before VAT) should be accompanied by remittance

Minimum transaction charge for cash order, regardless of the value of goods is £1.00

Our new 1978 Catalogue is now ready. Please send P.O. or stamps for 30p for your copy

WW-070 FOR FURTHER DETAILS

TELEPHONE TV IS HERE



THE ROBOT "530" PLTV TRANSCEIVER

ROBOT TELEPHONE LINE TELEVISION (PLTV) WILL SEND AND RECEIVE STILL TV PICTURES OVER ANY VOICE GRADE COMMUNICATIONS CHANNEL SUCH AS TELEPHONE OR RADIO. 2.0 KHZ is ample bandwidth and the picture is updated every eight seconds. This totally new concept is already proving invaluable for security, signature verification, surveillance, medical and other uses and is being used by Police, Military, Security firms, Meteorologists, Banks, Oil Gas and Electric companies, etc. etc.

The system is remarkably inexpensive and undoubtedly has many uses not as yet explored or envisaged.

If you think PLTV could be useful to you, or your clients, just drop us a line and we will send you full details.

Dealership opportunities exist in most U.K. areas and we will be pleased to hear from you if you have the facilities and ability to exploit this new concept of unlimited potential.

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WW-096 FOR FURTHER DETAILS

DRAKE Radio Shack Ltd



DRAKE SSR-1 GENERAL COVERAGE RECEIVER 0.5-30MHz 30 BANDS 10KHZ READOUT

£149.85 ACCESS OR BARCLAYCARD
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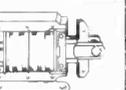
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10) N.M.S.

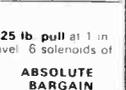
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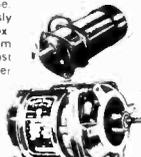


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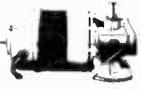


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Both types similar to above drawing. Price either type £4.75 + 75p
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Hewlett Packard	184A	DC-100MHz. Mainframe. Storage facility.	850.00
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	7A13	Dual Time Base. 2ns-5s/Div. Dual and delayed sweep. Triggering to 200MHz.	350.00
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	A100	Galvo 60Hz. 3.7µA/cm	450.00
	A2500	Galvo 1600Hz. 2.5mA/cm	30.00
Smiths	B450	Galvo 300Hz. 50µA/cm.	30.00
	RE551.20	X-Y Plotter A3. 50µV-500mV/cm sens on both axis. Timebase fitted	575.00
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Solartron	A200	19999 FSD DC only 1µV-1000V. 0.01% Autorange.	200.00
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NEW! TRANSCENDENT 2000

SINGLE BOARD SYNTHESIZER

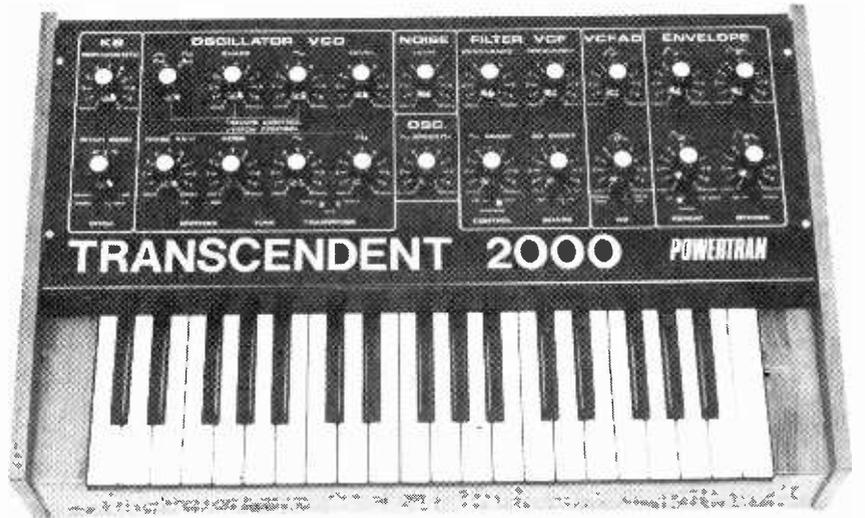
As featured in *Electronics Today International* (July, August 1978)

Live performance synthesizer designed by consultant Tim Orr (formerly synthesizer designer for EMS Ltd) and featured as a constructional article in *Electronics Today International*, the TRANSCENDENT 2000 is a 3 octave instrument transposable 2 octaves up or down with portamento, pitch bending, a VCO with shape modulation, a versatile VCF with both low and high pass outputs and a separate dynamic sweep control, a noise generator and an ADSR envelope shaper. There is also a slow oscillator and a new pitch detector amongst its many features.

Kit includes fully finished metalwork, solid teak cabinet, filter sweep pedal and really is complete — right down to the last nut and bolt. Virtually everything is on one circuit board and construction is so simple it can be built easily in a few evenings by almost anyone capable of neat soldering! When finished you will possess a synthesizer comparable in performance and quality with ready built units selling for between £500 and £700!

INTRODUCTORY OFFER
£172!

Due to the fantastic success of the launching of this superb new kit we are able to continue the Special Introductory Offer of £172 for complete kit



INTERNATIONAL POWERSLAVE 200+200 watt AMPLIFIER

As featured in *Electronics Today International*

400W rms continuous — 800W peak!
0.03% THD at FULL power!
PLUS all the following features too!

- * Each channel totally independent with its own stabilised power supply driven by custom designed TOROIDAL transformers!
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- * Value for money — quality and performance comparable with ready-built amplifiers costing over £600!

Pack	Price
1. Fibre glass printed circuit board for power amp	£4.20
2. Set of capacitors, metal oxide resistors, thermistor, cornet pre-sets for power amp	£6.40
3. Set of semiconductors for power and with mounting hardware, cooling tabs	£27.60
4. Pair of monster black drilled heat sinks, transistor mounting bracket	£6.90
5. Toroidal transformer, Primary 0-117V-234V, Secondaries 42-0-42V, 0-15V, 0-15V, Electrostatic screen	£17.50
6. Set of all parts for stabilized power supply including fibre glass printed circuit board, mounting bracket, semiconductors, resistors, capacitors, etc	£18.90
7A. Set of all parts for buffer/overdrive unit including fibre glass printed circuit board, semiconductors, resistors, capacitors, controls — required for PSI 4001 only	£3.80
7B. Set of parts for peak power meter including professional quality meter, fibre glass printed circuit board, components, controls — required for PSI 4002 only	£8.50
8. Set of all miscellaneous parts including sockets, illum. mains switches, fuse holders, fuses, cut-outs, cable, etc	£12.10
9. Cabinet, including chassis, anodised silver on black panels, fixing parts, etc. Please state whether Slave or Studio model required	£25.50
10. Handbook £0.50 or free on request when ordering any of above packs.	
2 each of packs 1-7 (A or B), 1 each 8, 9 and 10 are required for complete 200 + 200W professional amplifier.	
Total cost of individually purchased packs	PS1 4001 £208.20 PS1 4002 £217.60

PSI 4001 SLAVE MODEL



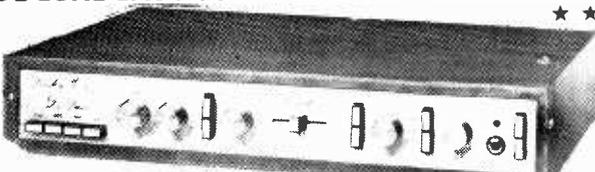
PSI 4002 STUDIO MODEL



SPECIAL PRICES FOR COMPLETE KITS!

PS1 4001 — £187.50
PS1 4002 — £196.90

DE LUXE EASY TO BUILD LINSLEY-HOOD 75W AMPLIFIER



AVAILABLE AS SEPARATE PACKS
PRICES IN OUR FREE CATALOGUE

SPECIAL PRICE FOR COMPLETE KIT

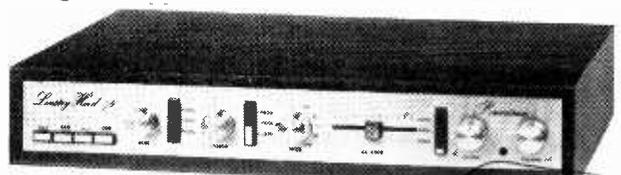
£99.30

The standard model of our kit for Mr Linsley Hood's 75 watt design has for a long time offered exceptional performance at a very modest cost with high quality high power ready-built units of comparable quality generally being over three times the price.

Features of the amplifier include very low distortion (less than 0.01%), 75W rms per channel power output, rumble filter, variable slope scratch filter, variable transition frequency tone controls, tape monitoring facilities and individually adjustable inputs. This model is based on 5 circuit boards which, not having the controls mounted on them, can, if desired, be effectively used separately in high performance audio systems not based on our metalwork.

Our new De Luxe model uses 14 boards which interconnect with gold plated contacts and are designed to have the potentiometers and switches mounted upon them. This system almost eliminates internal wiring, making installation after their assembly, delightfully straightforward, and as each board can be easily removed in seconds from the chassis, checking and maintenance is so simple that even newcomers to electronics will be able to cope competently with the kit. Additional features of our new model are inclusion of the latest circuit improvements, generously sized heat sinks for heavy duty use, even in tropical climates, and metal oxide resistors throughout for long-term stability and reliability.

STANDARD LINSLEY-HOOD 75W AMPLIFIER



SPECIAL PRICE FOR COMPLETE KIT

£79.80

Pack	Price
1. Fibreglass printed circuit board for power amp	£1.15
2. Set of resistors, capacitors, pre-sets for power amp	£2.50
3. Set of semiconductors for power amp	£6.50
4. Pair of 2 drilled, finned heat sinks	£1.10
5. Fibreglass printed-circuit board for pre-amp	£4.90
6. Set of low noise resistors, capacitors, pre-sets for pre-amp	£4.10
7. Set of low noise, high gain semiconductors for pre-amp	£2.40
8. Set of potentiometers (including mains switch)	£3.50
9. Set of 4 push-button switches, rotary mode switch	£5.40
10. Toroidal transformer complete with magnetic screen/ housing primary: 0 117-234 V; secondaries: 33-0-33 V, 25-0-25 V	£12.95

Pack	Price
11. Fibreglass printed-circuit board for power supply	£0.85
12. Set of resistors, capacitors, secondary fuses, semiconductors for power supply	£5.40
13. Set of miscellaneous parts including OIM skts., mains input skt., fuse holder, interconnecting cable, control knobs	£6.20
14. Set of metalwork parts including silk screen printed fascia panel and all brackets, fixing parts, etc.	£8.20
15. Handbook	£0.30
16. High Quality Teak Veneer cabinet 18.3" x 12.7" x 3.1"	£10.70

2 each of packs 1-7, 1 each of packs 8-16 inclusive are required for complete stereo amplifier. Total cost of individually purchased packs £92.80

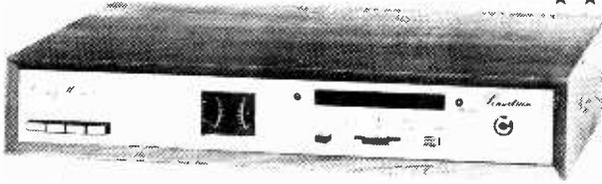
PACK PRICES FOR STANDARD KIT

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MALTA

Designed in response to demand for a tuner to complement the world-wide acclaimed Linsley-Hood 75W Amplifier, this kit provides the perfect match. The Wireless World (Skingley and Thompson) published original circuit has been developed further for inclusion into this outstanding slimline unit and features a pre-aligned front end module, excellent a.m. rejection and temperature compensated varicap tuning, which may be controlled either continuously or by push-button pre-selection. Frequencies are indicated by a frequency meter and sliding LED indicators, attached to each channel selector pre-set. The PLL stereo decoder incorporates active filters for "birdy" suppression and power is supplied via a toroidal transformer and integrated regulator. For long term stability metal oxide resistors are used throughout.

AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

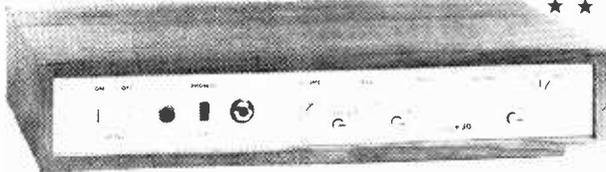
LINSLEY-HOOD CASSETTE DECK



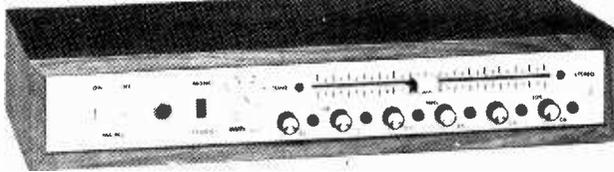
SPECIAL PRICE FOR COMPLETE KIT £79.60

Published in Wireless World (May, June, August 1976) by Mr. Linsley-Hood, this design, although straightforward and relatively low cost, nevertheless provides a very high standard of performance. To permit circuit optimization separate record and replay amplifiers are used, the latter using a discrete component front-end designed such that the noise level is below that of the tape background. Pushbutton switches are used to provide a choice of equalization time constants, a choice of bias levels and also an option of using an additional pre-amplifier for microphone use. The mechanism used is the Goldring-Lenco CRV, a unit distinguished in its robustness and ease of operation. Speed control and automatic cassette ejection are both implemented by electronic circuitry. This unit which is powered by a toroidal transformer and uses metal oxide resistors throughout offers an excellent match for the Wireless World Tuner and the Linsley-Hood 75 Watt Amplifier. Circuit changes as published in February, 1978, follow-up article are included in the kit AT NO EXTRA COST! A higher performance head (Matsushita WY 436 AZ head as recommended in the follow-up article) is offered as an optional extra but this will be automatically supplied FREE OF CHARGE with all orders for complete kits!

**T20 + 20 AND T30 + 30
20W, 30W AMPLIFIERS**



WWII TUNER



SPECIAL PRICE FOR COMPLETE KIT £47.70

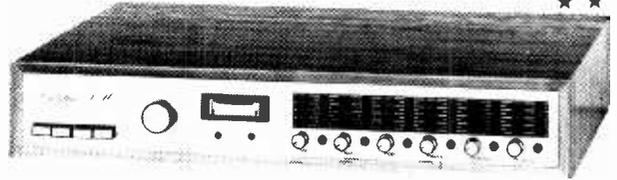
AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

Following the success of our Wireless World FM Tuner Kit this cost reduced model was designed to complement the T20+20 and T30+30 amplifiers and the cabinet size, front panel format and electrical characteristics make this tuner compatible with either

Wireless World Designs: Full kits are not available for the projects below but PCBs and component sets are stocked. Further details of these and other packs are in our Free Catalogue

30W Bailey Amplifier		Regulated Power Supply for Bailey Amplifier	
BAIL Pk 1 F Glass PCB	£1.00	TRRC Pk 1 F Glass PCB	£0.85
BAIL Pk 2 Resistors Capacitors	£2.35	60VS Pk 2 Resistors Capacitors	£2.20
BAIL Pk 3 Semiconductors	£4.70	60VS Pk 3 Semiconductors	£3.10
		60VS Pk 6A Toroidal transformer	£8.80
Linsley-Hood Low Distortion Oscillator.		Stuart Tape Recorder	
LDO Pk 1 Fibreglass PCB	£1.65	TRRC Pk 1 Replay Amp F G PCB (stereo)	£1.30
LDO Pk 2 M.O. Resistors capacitors	£2.60	TRRC Pk 2 Record Amp F G PCB (stereo)	£1.70
LDO Pk 3 Semiconductors	£3.90	TROS Pk 1 Bus Erase F G PCB (stereo)	£1.20
E. F. Taylor Pre-Amplifier			
EFTF Pk 1 Fibreglass PCB (stereo)	£1.45		
EFTF Pk 2 M.O. Res. caps (stereo)	£3.20		
EFTF Pk 3 Semiconductors (stereo)	£4.20		

WIRELESS WORLD FM TUNER



SPECIAL PRICE FOR COMPLETE KIT £70.20

Pack 1	Stereo PCB (accommodates 2 rep. amps. 2 meter amps. bias/erase esc. relay)	£3.35	Pack 10	Set of capacitors, rectifiers, I.C. voltage regulator P.C.B. for power supply (Powertran design)	£2.80
Pack 2	Stereo set of capacitors, M.O. resistors, potentiometers for above	£7.95	Pack 11	Set of miscellaneous parts, including sockets, fuse holder, fuses, interconnecting wire, etc.	£3.40
Pack 3	Stereo set of semiconductors for above	£8.50	Pack 12	Set of metalwork including silk screened fascia panel, internal screen, fixing parts, etc.	£7.10
Pack 4	Miniature relay with socket	£2.90	Pack 13	Construction notes	£0.25
Pack 5	PCB, all components for solenoid, speed control circuits	£3.80	Pack 14	High Quality Teak Veneer cabinet 18.3" x 12.7" x 3.1"	£10.70
Pack 6	Goldring-Lenco mechanism as specified	£18.50			
Pack 7	Function switch, knobs	£1.90			
Pack 8	Dual VU meter with illuminating lamp	£6.95			
Pack 9	Toroidal transformer with E.S. screen prim. 0-117V. 234V. Sec. 15V	£4.90			

One each of packs 1-14 inclusive are required for complete stereo cassette deck. Total cost of individually purchased packs £83.00

Matsushita WY 436 AZ head (optional extra) £4.50 (free with compete kit)

Designed by Texas engineers and described in Practical Wireless, the Texan was an immediate success. Now developed further in our laboratories to include a Toroidal transformer and additional improvements, the slimline T20 + 20 delivers 20W rms per channel of true Hi Fi at exceptionally low cost. The easy to build design is based on a single F Glass PCB and features all the normal facilities found on quality amplifiers including scratch and rumble filters, adaptable input selector and headphones socket. In a follow up article in Practical Wireless further modifications were suggested and these have been incorporated into the T30 + 30. These include RF interference filters and a tape monitor facility. Power output of this model is 30W rms per channel.

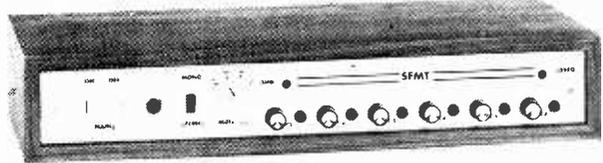
SPECIAL PRICES FOR COMPLETE KITS

T20+20 KIT PRICE £33.10

T30+30 KIT PRICE £38.40

AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

POWERTRAN SFMT TUNER



PRICE FOR COMPLETE KIT £35.90

AVAILABLE AS COMPLETE KIT ONLY

This is a simple, low cost design which can be constructed easily without special alignment equipment but which still gives a first-class output suitable for feeding any of our popular amplifiers or any other high quality audio equipment. A phase-locked-loop is used for stereo decoding and controls include switchable afc, switchable muting and push-button channel selection (adjustable by controls on the front panel). This unit matches well with the T20 + 20 and T30 + 30 amplifiers.

NEW! Improved stereo decoder (as described in April 1978 W W) F/Glass PCB, M.O. Res. Caps. Cermet pre-sets, IC IC socket. **£6.30**

SQ QUADRAPHONIC DECODERS

These state-of-the-art circuits described by CBS are offered as kits of superior quality with close tolerance capacitors, metal oxide resistors and Fibreglass PCBs designed for edge connector insertion. Further information on these kits is given in our FREE CATALOGUE

M1 Basic matrix decoder	£5.90
L1 Full logic decoder	£17.20
L2A Full logic decoder with variable blend	£22.60
L3A As L2A but with high performance discrete component front end	£30.10
(or with carbon film resistors)	£25.90
SQM1-30 Decoder complete with 30W rear channel amplifiers. Complete kit matches T30 + 30 amplifier	£40.75

EXPORT A SPECIALITY!

Our Export Department can readily despatch orders of any size to any country in the world. Some of the countries to which we sent kits last year are shown in this advertisement. To assist in estimating postal costs our catalogue gives the weights of all packs and kits. This will be sent free on request by airmail together with our "Export Postal Guide" which gives current postage prices. There is no minimum order charge. Prices the same as for U.K. customers but no Value Added Tax charged. Postage charged at actual cost plus 50p documentation and handling. Please send payment with order by Bank Draft, Postal Order, International Money Order or cheque drawn on an account in the U.K. Alternatively for orders over £500 we will accept Irrevocable Letter of Credit payable at sight in London.

Value Added Tax not included in prices UK Carriage FREE

SERVING FACILITIES: Available for all ★★ complete kits. **PRICE STABILITY:** Order with confidence! Irrespective of any price changes we will honour all prices in this advertisement until Sept 30th 1978 if this month's advertisement is mentioned with your order. Errors and VAT rate changes excluded.

U.K. ORDERS: Subject to 12½% surcharge for VAT (i.e. add ½ to the price). No charge is made for carriage or current rate if charged. **SECURICOR DELIVERY:** For this optional service (U.K. mainland only) add £2.50 (VAT inclusive) per kit. **SALES COUNTER:** If you prefer to collect your kit from the factory call at Sales Counter (at rear of factory) Open 9 a.m. - 4.30 p.m. Monday-Thursday.

QUALITY: All components are brand new first grade full specification guaranteed devices. All resistors (except where stated as metal oxide) are low noise carbon film types. All printed circuit boards are fibreglass drilled roller tinned and supplied with circuit diagrams and construction layouts.

FOR FURTHER INFORMATION PLEASE WRITE OR TELEPHONE FOR OUR FREE CATALOGUE

POWERTRAN ELECTRONICS
PORTWAY INDUSTRIAL ESTATE ANDOVER
ANDOVER HANTS SP10 3NN (0264) 64455

DEPT. WW8

ADVANCE BOARD No. 70065. A modern fibre glass circuit board made for computer but never issued, all components can be assumed perfect.
Major items — 8 transistors type BC 107, 8 transistors type BC 212, 9 miniature diodes, 4 preset variable pots, 2 200UF 63v capacitor, 2 47UF 40 volt capacitors, 2 220UF 10 volt capacitors, 1 1UF 63 volt capacitor, 50 assorted resistors 1/4, 1/2 watt. Board size approx. 5 x 4 1/2". Most components can be removed with working length leads **£1.00.**

Telephone Answering Machine. Used, but we understand are in good working order, however, we can supply only for breaking up — they are not to be used for telephone answering. They contain

4 pole tape motor, twin capstans with heavy flywheels, 4 record-playback heads, erase head, tape spools to take standard reels, tape guides and solenoid operated brake, 8 piano key type switching and control mechanism, tape used counter, elliptical speaker, 2 solenoids, 9 pin plug and socket, 36 pin plug and socket 5 Circuit boards containing varied assortment of transistors in small parts. All the above mentioned components are mounted in the main chassis and there is a sub chassis with 5 miniature 4 pole relays, ferrite transformer, 4 iron cored transformers, 5 variable pots, 35 transistors, over 300 various resistors, capacitors, diodes, etc., full wave rectifier panel, 4 way push down wire connection block, two-way ditto, stereo input socket, 3 push-in neon bulbs, lid switch, 4 electrolytic capacitors, 2 power output transistors. The unit is nicely cased size about 27" x 10" 6" and should be suitable for conversion to open reel tape recorder, background music machine, echo chamber etc. Price **£12.25.**

High Voltage Mains Transformer. Normal mains primary-secondary by our measuring equipment is 8KV approx at 5 mA. We are offering these at a bargain price of **£4.75.** Our ref. no. TM45.

Sorry, sold out of transformer ref. No. TM 37 but we have just received another transformer which may fill your need. This is a 100v transformer 80 volt secondary, tapped at 20, 40 and 60 volts. So this could be used as a 80v, 1.5 amp, 60v at 2 amp, 40 volt 3 amp, a 20-0-20 volt 3 amp or a 40-0-40v 1.5 amp, price **£5.55.**

Smith's Blower. Snail shape with exterior motor, oblong outlet size 4 1/2" x 1 1/4" approx., paddle type air rotor coupled to mains induction motor, with anti vibration mountings overall size of fan approx 4 1/2" high x 6 1/2" wide x 4 1/2" diameter. **£5.80.**

Torin Blower. Snail type similar but smaller to above, aperture size 2 1/2" x 1 1/4" approx. Normal mains induction motor overall size 5 1/2" high x 5 1/4" wide x 5 1/2" diameter. Price **£5.86.**

High Voltage Capacitors. 80 of 5 KV working eq equipment but with useable length leads. Normally a very expensive capacitor, our price **25p + 2p each.**

Flex Bargain. 3 Core (standard colour coding), black outer pvc cover, all made from heat resistant plastic. Suitable for connecting direct to heating appliances but being tougher than usual it is ideal also for extension leads especially outdoor ones. 5mm conductors so suitable for up to 7.5 amp 100 met. coil, price **£10.95.**

1/2 rev per minute mains driven motor, 2 watts also suitable only for timers or other lightweight operations, price **£2.70.**

Moving Coil Panel Meter. 75 mA fsd 2 1/2" diameter this is an experiment item and the scale is headed Radiation, limited quantity available **£1.89.**

A Clock Switch, 12 hour type as fitted to cookers. This comprises a normal 12 hour continuously running clock coupled to 25 amp switches with levers for setting time plus a further minute winder bell switchable up to 60 minutes. Less knobs which can be quite easily made from plastic rod. Size approx. 7 1/4" wide, 3 1/4" high, 3" deep, limited quantity **£2.12.**

Mains Relay single screw fixing open type with single 10 amp changeover contact eq equipment with our usual guarantee price **60p + 4p, 10 for £5 + 40p.**

Adjustable Air Thermostat with 15 amp contacts, spindle protrudes enough for normal type knob, can be easily assessable, screw enables these to be set for normal air temperatures 30-80 for lower or higher. **50p + 4p each, 10 for £4.50 + 36p.**

Car Speakers. Two bargains this month both elliptical, both 4 ohm size. Price **£1.50 + 19p, post 60p.**

Immersion Heater Thermostats, made by Satchwell 7", 11" and 17" lengths, standard fit in most immersion heaters. **£2.10.**

Thermostat Pocket, to fit the above thermostat into a tank without heater then you need a pocket to hold the thermostat. 18" long threaded complete with nut and washer, price **£1.62.** Also available 8" long for 7" thermostat, same price.

Hot Wire Vacuum Relay Switch, 4 pin plug-in type. This has a heater coil wound around a bi-metallic strip causes switch on after a time which is adjustable. The energising voltage varies between 4v & 7v time delay time — from a few seconds upwards. Price **£1.08.**

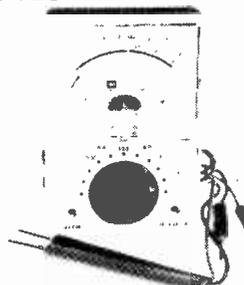
Boiler Stat, 60-90 C remote phial type capillary length approx. with control knob marked 20 to 80 C, price **£2.16.**

Transistor holder for 105°C 26, etc.) allows transistor to be replaced quickly, also threaded for holding screws. AC capacitor 1.25µF for 40 volts rms. Aluminium can with tag connections eq new equipment covered by normal guarantee. **30p + 2p.**

MINI-MULTI-TESTER

Mazing little pocket size precision moving coil instrument — jewelled bearings — 1000 opv — mirrored scale — 11 instant ranges measures

- DC volts 10 50 250 1000
- AC volts 10 50 150 1000
- DC amps 0.1 mA and 0.100 mA
- Continuity and resistance 0.150K ohms
- Complete with insulated probes, leads, battery, circuit diagrams and instructions
- Unbelievable value only **£6.50.**



MULLARD UNILEX

A main powered 4 + 4 stereo system rated one of the finest performers in the stereo field this would make a wonderful gift for almost anyone in easy-to-assemble modular form and complete with a pair of Plessey Speakers this should sell at about £30 — but due to a special bulk buy and as an incentive for you to buy this month we offer the system complete at only **£15** including VAT and postage. 10 watt amps to upgrade unilex **£3.50 each**



UNISELECTORS

These are pulse operated switches as used in automatic telephone switchboards etc. The pulse moves the switch arm through one position. Except where indicated the selectors are 25 position types and 50v. Coil is standard 24v or 12v operation extra at **£2** per switch

3 pole	£5.90
4 pole	£6.98
5 pole	£8.20
6 pole	£9.25
8 pole	£11.42
10 pole	£13.60
12 pole	£15.88
2 pole 50 way	£8.60
3 pole 50 way	£11.40

24 HOUR TIMERS

The one illustrated is the E control this uses the Smiths mechanism as in their autoser. 2 on off s per 24 hours 13 amp contacts override switch **£6.50.**
Smiths 100 amp model one on off per 24 hours **£10.50**, extra contacts **£1.00** per set
AEG 60 amp model with clockwork standby, one on off per 24 hours **£9.50**, extra contacts **£1.00** per set

RELAYS

12 volt two 10 amp changeover plug in **95p** 12v three 10 amp changeover plug in **£1.28.** 12v two changeover miniature wire ended **85p.** 12 volt open single screw fixing two 10 amp changeovers **85p** 12 volt open three 10 amp changeovers **£1.25.** Latching relay mains operated 2 c o contacts **£2.11.** Mains operated three 10 amp changeovers open type one screw fixing **£1.25.** Many other types with different coil voltages and contact arrangements are in stock enquiries invited



ROTARY PUMP

Self priming portable, fits drill or electric motor, pumps up to 200 gallons per hour depending upon revs. Virtually uncorrodible, use to suck water of petrol/fertiliser chemicals, anyting liquid. Hose connectors each end. **£2** post paid



DELAY SWITCH

Mains operated — delay can be accurately set with pointers knob for periods of up to 2 1/2 hrs 2 contacts suitable to switch 10 amps — second contact opens few minutes after 1st contact **95p.**

HUMIDITY SWITCH

American made by Rank, the type No. 211. The action of this device depends upon the dampness causing a membrane to stretch and trigger a sensitive microswitch adjustable by a screw quite sensitive — breathing on it for instance will switch it on. Micro 3 amp at 250V ac. Overall size of device approx 3 1/2" long 1 1/2" wide and 1 1/2" deep. **75p.**



INDUCTION MOTORS

One illustrated is our reference MM11 made for ITT 1/4" stack 1 1/2" spindle **£2.25,** 1/2" stack model **£1.75,** 1" stack **£2.75,** 1 1/2" stack **£3.25.**



EXTRACTOR FAN

Ex-computers made by Woods of Colchester, ideal for fixing through panel — reasonably quiet running — very powerful 25000 rpm. Choice of two sizes 5 or 6 1/2" dia **£5** and **£6.**

SMITHS CENTRAL HEATING CONTROLLER

Push button gives 10 variations as follows: (1) continuous hot water and continuous central heating (2) continuous hot water but central heating off at night (3) continuous hot water but central heating on only for 2 periods during the day (4) hot water and central heating both on but day time only (5) hot water all day but central heating only for 2 periods during the day (6) hot water and central heating on for 2 periods during the day time only — then for summer time use with central heating off (7) hot water continuous (8) hot water day time only (9) hot water twice daily (10) everything off. A handsome looking unit with 24 hour movement and the switches and other parts necessary to select the desired programme of heating. Supplied complete with wiring diagram. Originally sold we believe at over **£15** — we offer these while stocks last at **£7.50** each including VAT & Postage



TERMS: Prices include Post & VAT, but orders under £6.00 please add 50p to offset packing etc. Bulk enquiries please phone for generous discount 01-688 1833

J. BULL (ELECTRICAL) LTD
(Dept. WW)
103 TAMWORTH ROAD
CROYDON CR9 1SG

IT'S FREE!

Our monthly Advance Advertising Bargains List gives details of bargains arriving or just arrived — often bargains which sell out before our advertisement can appear. — It's an interesting list and it's free — just send S.A.E. Below are a few of the Bargains still available from previous lists.

Mains operated Siren. Don't let intruders get away with your possessions — they will never stay in a house when one of these sirens is going. Quite small but very alarming **£13.50.**

Lever Switch as fitted to modern telephone switchboards. 8 pole changeover contacts made by Pye/TMC pushed to return when pushed up, stays down when pushed down, price **£1.08.**

Pulsing Switch. Motorized unit which gives pulses every 30 seconds, length of pulse can be adjusted up to 30 seconds and the pulse can be up to 20 amps at normal mains voltage. Made up by famous Cramer Company of America, the drive motor of this device is 115v 50hz but we supply complete with series voltage dropping device to make it suitable for our mains. This is in a cylindrical plastic case overall size, with a knob on the front for adjusting the pulse length. 20 amp switch 'inside' is a changeover switch so this device could also be used as a time sharing switch, when one circuit is on the other circuit would be off for a length of time determined by switch control setting, price **£4.82.**

16 Line Connecting Box. This is 16 way twin grub screw type connecting strip, mounted in a standard 2 gang MK white surface box with cover made for Satchwell so obviously a good product. The cables, are brought in through breakaways in the plastic box. 16 connection points are all numbered for easy identification, price **£1.92.**

Air Thermostat with remote setting dial. This is a Satchwell thermostat using a sensor connected to the switch by a 26" length of capillary. The control setting adjustable from 30 to 140 F complete with control knob showing temperature setting **£2.46.**

Twin 13 amp Rocker Switches (DOT), price **49p** the pair. **Pressure Gauge,** standard airline thread, Reads 0-30 lbs per sq inch. Price **81p.**

Bargain for callers only. VDU with 18" CR Tube rather large **£14.50.**

4 Way Terminal Blocks, twin grub screw type. PVC covered 10 for **65p.**

Spares for Dimplex Heaters. We have just taken delivery of a large quantity of various spare parts for Dimplex heaters including storage heaters, if you need any of these then please let us have your enquiries

Heavy Duty Casters. Four of these would carry a ton, set of 4 **£2.65.**

Super Power 2N3055, RCA 52360, in our trials this does all that the 2N3055 will do but very much better. Truly a remarkable transistor **85p.**

200 watt Transformer, 40v-0-40v, normal type construction and primary wound for 230 50hz, **£6.65.**

High Voltage Rectifier, 5kv working at 5mA these are unused equipment but have good length leads, ideal for use with the EHT transformer if joined in series, price **33p.**

Speaker Cabinets, Simulated teak finish, nice handy size, modern black sponge type front **£3.75.**

In Car Speaker Cabinet, White with black edge very modern looking plastic with threaded studs for mounting speaker with back, price **£2.25.**

AC Capacitors for use on fluorescent lighting for power factor correction or as a voltage dropping device, these are very rugged and will stand DC voltages up to 3 times their RMS voltage. A big purchase enables us to offer these at about one-third of the current manufacturers price, all are 300v RMS working or higher and are in aluminium cans with tags or wire ends, following values 1 25µF **38p,** 1 5µF **49p,** 3µF **59p,** 4µF **70p,** 7µF **92p,** 8µF **£1.07,** 15µF **£1.35.**

Numerical Display Tubes (Nixie tubes), Mullard ref. ZM 1175, this is a sideways viewing device which displays all figures from 0-9, has wire leads, now in box **92p** each

Waterproof Diecast Box, very suitable for protecting a switch or a thermostat or a similar device where this is mounted outside or in a greenhouse, price **£1.62.**

Multiview Switches, GEC silver finished metal box with cable knockouts each complete with switch mounting grid and matching recessed cover, suitable for conduit or TRS. Single switch **50p,** twin switch **60p,** 4 switch **75p,** 6 switch **£1.12,** 12 switch **£1.50.**

Modern toggle type miniature switches by GEC to fit above boxes, mains rating 5 amp on/off **35p,** 15 amp on/off **45p,** 5 amp 2 way **30p,** 2 way and off **50p,** intermediate (polarity changeover **50p),** bell push **35p** (available in several colours. Please add 8% VAT to total cost of boxes and switches

Most of the above switches can be supplied without toggles but operated by a special key, add **10p** per switch and **25p** per key

Can Any Reader help! We have recently acquired some very nice American made motors 50 cycle for 50hz 220v working obviously made for the British Market but they have 5 lead out wires and we have not been able to find out the correct method of working. It is possible that they need a capacitor. The colours of the leadout wire are red, white, yellow and blue. The maker's name is Robbins and Myers and the model number of the motor is KS-PP30-601 rated at 1/12 hp single phase 1425 rpm. Price of the motor **£5.50 + 44p,** Post and packing **£2.**

Boiler Stat, Satchwell remote dial type with knob calibrated 20-90 C. Price **£2.42.**



MINIATURE RELAY

12v dc operated with two sets of changeover contacts. The unique feature of this relay is its heavy lead out wires. These provide adequate support and therefore the relay needs no fixing. On the other hand there is a fixing bolt through one side so if you wish you can fix the relay and use its very strong lead outs to secure circuit components—an expensive relay. We are offering them at only **87p** each

Don't miss this exceptional bargain.

T.T.L. 74 I.C.s By TEXAS, NATIONAL, I.T.T., FAIRCHILD ETC.

7400 14p	7426 25p	7473 30p	74121 30p	74151 65p	74179 140p
7401 14p	7427 25p	7474 30p	74122 40p	74153 65p	74180 100p
7402 14p	7428 25p	7475 30p	74123 60p	74154 120p	74181 200p
7403 14p	7429 25p	7476 40p	74125 50p	74155 70p	74182 75p
7404 14p	7430 25p	7477 40p	74126 50p	74156 70p	74184 150p
7405 14p	7431 25p	7478 40p	74130 100p	74157 70p	74185 150p
7406 40p	7432 25p	7479 40p	74131 100p	74160 90p	74188 350p
7407 40p	7440 15p	7489 250p	74132 65p	74161 90p	74189 350p
7408 20p	7441 15p	7490 35p	74135 100p	74162 90p	74190 350p
7409 20p	7442 65p	7491 75p	74136 80p	74163 90p	74191 140p
7410 15p	7443 30p	7492 45p	74137 100p	74164 125p	74192 120p
7400 20p	7446 80p	7493 40p	74138 125p	74165 125p	74193 120p
7412 20p	7447 75p	7495 60p	74139 100p	74166 125p	74194 100p
7413 30p	7448 70p	7496 70p	74141 60p	74167 325p	74195 100p
7414 60p	7450 15p	74100 95p	74142 270p	74170 200p	74196 100p
7415 30p	7451 15p	74104 90p	74143 270p	74173 150p	74197 100p
7416 30p	7452 15p	74105 40p	74144 270p	74174 100p	74198 185p
7417 30p	7453 15p	74107 30p	74145 75p	74175 75p	74199 185p
7418 30p	7454 15p	74109 50p	74147 230p	74176 100p	
7419 30p	7455 15p	74110 90p	74148 160p	74177 100p	
7420 25p	7456 15p	74111 90p	74149 120p	74178 140p	
7421 25p	7457 15p	74112 90p			

C.MOS

4000 14p	4030 55p
4001 14p	4032 55p
4002 14p	4033 120p
4006 90p	4047 100p
4007 15p	4048 55p
4009 55p	4049 40p
4011 14p	4040 40p
4012 14p	4054 120p
4013 50p	4055 140p
4015 90p	4056 135p
4016 40p	4060 120p
4017 90p	4066 55p
4018 90p	4069 20p
4020 100p	4071 16p
4022 90p	4072 15p
4023 16p	4081 15p
4024 55p	4082 16p
4025 16p	4510 120p
4026 160p	4511 150p
4027 50p	4516 110p
4028 90p	4518 130p
4029 110p	4528 100p

IN1418 BY ITT/TEXAS 100 for **£1.50**. FULL SPEC

UNENCODED HEXADECIMAL 19 KEYBOARD 0-9 A B C D E F 2 OPTIONAL KEYS SHIFT KEY £12.50

555 Timer, 10 for **£2.50**.

741 Op amp 10 for **£2.00**.

RCA SCR to 3 case 100V 12.5A **£2.50**.

MURATA ULTRASONIC TRANSDUCERS MA40WR/S **£2.50** each, £4.00 pair.

210AN-2L 1024 x 1 BIT 250 NANO SEC. STATIC RAM **£2.60**, 47 **£8.40**, 8 **£16.00**.

2 102AN-4L 1024 x 1 BIT 450 NANO SEC. STATIC RAM **£2.60**, 4/ **£6.00**, 8/ **£11.60**.

2112-4 256 x 4 BIT 450 NANO SEC. STATIC RAM **£2.95** each 4 **£11.60**, 8/ **£22.80**.

2513 CHARACTER GENERATOR UPPER CASE **£7.00**.

2513 CHARACTER GENERATOR LOWER CASE **£7.00**.

MM5204AQ PROM 4096 BIT READ ONLY MEMORY **£8.00**.

B212 8 BIT IN/OUT PORT **£3.00**.

8831 TRI-STATE QUAD LINE DRIVER **£2.00**.

8833 TRI-STATE TRANSCIEVER (TRUE) **£2.00**.

8835 TRI-STATE TRANSCIEVER (INVERTING) **£2.00**.

AY5-1013 UAR/T **£6.00**.

LM309K/LM3340K VOLTAGE REGULATOR **£1.00** each.

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240v Plug-In Relays 3 pole c o 10 amp contacts 85p P.P. 15p.

COAXIAL CRYSTAL DETECTORS. (Marconi-Saunders). 200 MHZ-12 GHZ £7.50 (Diode CS9B £1.50)

FIBREGLASS COPPER-CLAD BOARD
9x 4 1/2 x 1/16in 40p P&P 10p
9 x 6 x 1/16in 50p P&P 15p

OFF-CUT PACKS. 150 sq ins. £1 P.P. 25p
Double sided 1/2p per sq. in. extra

MAGNETIC COUNTERS
3 Digit Reset (240v A.V) £1.75 P&P 25p
6 Digit Reset (24v D.C.) £4 P&P 25p
4 Digit Non-Reset (24v D.C.) £1 P&P 25p
6 Digit Non-Reset (240v A.C.) £1.50 P&P 25p

MULTICORE CABLES
4 CORE RIBBON (RAINBOW) CABLE 4 - 10/ 2m m.
Forming 1/4in wide strip. 10m-75p 50m-£3 100m-£6 P&P 1p per metre
8 CORE RIBBON (RAINBOW) CABLE 8 x 14/76
Forming 1/2in wide strip. 10m-£1.50 50m-£6.50 100m-£12. P&P 1p per metre
10 CORE CABLE 10 x 7/76 (10 colours) P.V.C
O.D. 7m m 10m-£2 50m-£8.50 100m-£16 P&P 2p per metre
12 CORE SCREENED CABLE 12 x 14/76 with outer screen - P.V.C covered O.D. 9m m.
10m-£4 50m-£18.50 100m-£35 P&P 2p per metre
16 PAIR RIBBON CABLE 16 x 2 core P.V.C
Double sheathed forming 2in wide strip
10m-£3.50-£13.50 100m-£25 P&P 2p per metre

E.H.T. MODULES (Resin encapsulated in metal box)
Input 240v 50hz Type 1 O/P 8kv @ 15 watts £75 P&P £1
Type 2 O/P 13.7kv @ 7 watts £10 P&P £1

STABILISED POWER SUPPLY (APT) 5-14 volts @ 6-amp Pre-Set (with manual) £20 P.P. £2

P.C. EDGE CONNECTORS
32 way (1 pitch) finished ends 40p P&P 10p
56 way (1 pitch) cuttable 65p P&P 15p
64 way (1 pitch) cuttable 75p P&P 15p
64 way gold plated pins 90p P&P 15p
Mounting pillars for 56/64 way 15p per pair

'DRYFIT' RE-CHARGABLE BATTERIES (Lead/Acid)
Ex Equip Good condition tested
6v @ 6 A.H. £3.50 P&P 75p
6v @ 7.5 A.H. £4.00 P&P 75p

'BLEEPSTONE' AUDIO ALARMS (Ex Equip) 12v D.C. 75p P.P. 10p

EMERGENCY LIGHTING UNITS. Automatically Switches to stand-by-battery power (24v) when mains fails. Keeps battery pack fully charged on normal supply. £15 (ex batteries) P.P. £1.50.

VARLEY LEAD-ACID BATTERIES. 6v 36 A.H. non-spill £6 ea P.P. £1.50

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TRANSFORMERS
SCREENED - SAME-DAY DESPATCH
MAINS ISOLATING VAT 8% 12 and/or 24-VOLT
PRI 120/240V SEC 120-240V
Centre Tapped and Screened

Ref. VA (Watts)	£	P&P	Ref	12v	24v	£	P&P
07* 20	4.40	79	111	0.5	0.25	2.20	.45
149 60	6.20	96	213	1.0	0.5	2.64	.78
150 100	7.13	114	71	2	1	3.51	.78
151 200	11.16	114	18	4	2	4.03	.96
152 250	12.79	150	70	6	3	5.35	.96
153 350	16.28	184	108	8	4	6.98	1.14
154 500	19.15	215	72	10	5	7.67	1.14
155 750	29.06	OA	116	12	6	8.99	1.32
156 1000	37.20	OA	17	16	8	10.39	1.32
157 1500	45.60	OA	115	20	10	13.18	2.08
158 2000	54.80	OA	187	30	15	17.05	2.08
159 3000	79.05	OA	226	60	30	26.82	OA

*115 or 240 sec only State volts required.

30 VOLT RANGE
Pri 220-240V
Sec. 0-12-15-20-24-30V
Volts available 3 4 5 6 8 9 10 12 15 18 20 24 30V or 12V-0-12V and 15V-0-15V

Ref.	Amps	£	P&P
112	0.5	2.64	.78
79	1.0	3.57	.96
3	2.0	5.27	.96
20	3.0	6.20	1.14
21	4.0	7.44	1.14
51	5.0	8.37	1.32
117	6.0	9.92	1.45
88	8.0	11.73	1.64
89	10.0	13.33	1.84

50 VOLT RANGE
Pri 220-240V Sec. 0-20-25-33-40-50V
Volts available 5 7 8 10 13 15 17 20 25 30 33 40 or 20V-0-20V and 25V-0-25V Screened

Ref.	Amps	£	P&P
102	0.5	3.41	.78
103	1.0	4.57	.96
104	2.0	6.98	1.14
105	3.0	8.45	1.32
106	4.0	10.70	1.50
107	6.0	14.62	1.64
118	8.0	17.05	2.08
119	10.0	21.70	OA

60 VOLT RANGE
Pri 220-240V Sec. 0-24-30-40-48-60V Volts available 6 8 10 12 16 18 20 24 30 36 40 48 60V or 24V-0-24V and 30V-0-30V

Ref.	Amps	£	P&P
124	0.5	3.88	.96
126	1.0	5.58	.96
127	2.0	7.60	1.14
125	3.0	10.54	1.32
123	4.0	12.23	1.84
40	5.0	13.95	1.64
120	6.0	15.66	1.84
121	8.0	20.15	OA
122	10.0	24.03	OA
189	12.0	27.13	OA

HIGH VOLTAGE
MAINS ISOLATING
Pri 200/220 or 400/440
Sec 100/120 or 200/240

VA	Ref.	£	P&P
60	243	5.89	1.32
350	247	14.11	1.84
1000	250	41.76	OA
2000	252	54.25	OA

BRIDGE RECTIFIERS

200v	2A	45p
400v	2A	55p
200v	4A	65p
400v	4A	80p
400v	6A	£1.05
500v	10A*	£2.35

*P&P 15p. VAT 12 1/2% *VAT 8%

TEST METERS

AV08 Mk. 5	£77.10
AV071	£31.30
AV073	£42.50
AV0MM5 MINOR	£26.10
WEE MEGGER	£64.00

AVO TT169 (tests transistors in circuit no soldering) **£32.50**
UA4315 budget meter (42 ranges) 20kV V/DC 1000V AC/DC (9 ranges) 2.5A AC/DC 500K resistance in robust steel case with leads, full instructions. **£15.85**
EM272 **£50.70**
DA116 Digital Megger BM7 (Battery) **£42.50**
Avo Cases and Accessories P&P £1.15 VAT 8%

MINI-MULTIMETER
DC1000V AC-1000V AC/DC-1000V/V DC-100mA Res - 150K Bargain at **£5.86** VAT 8% P&P 62p

ABS PLASTIC BOXES
Inset brass units. slots to take PC cards (boards) flush fitting lid.
PB1 80mm x 62 x 40 **.56p**
PB2 100mm x 75 x 40 **.63p**
PB3 120mm x 100 x 45 **.70p**
P&P 29p VAT 8%

SPECIAL OFFER
Avo Meter (107) Ranges DC to 1000V current to 1 amp AC 1000V 3 Res 3 amp ranges to 10M ohms 10 000 DC 1 000 AC safety cut out **£40.00**

SCREENED MINIATURES Primary 240V

Ref.	mA	Volts	£	P&P
238	200	3-0.3	1.99	.55
212	1A 1A	0-6 0-6	2.85	.78
13	100	9-0.9	2.14	.38
235	330 330	0-9 0-9	1.99	.38
207	500 500	0-8-9 0-8-9	2.59	.71
208	1A 1A	0-8-9 0-8-9	3.53	.78
236	200 200	0-15 0-15	1.99	.38
239	50MA	12-0-12	1.99	.38
214	300 300	0-20 0-20	2.56	.78
221	700 (DC)	20 12-0-12-20	3.41	.78
206	1A 1A	0-15-20 0-15-20	4.63	.96
203	500 500	0-15-27 0-15-27	3.99	.96
204	1A 1A	0-15-27 0-15-27	5.39	.96
S112	500	0-12-15-20-24-30	2.64	.78

Step Up or Step Down

CASED AUTO. TRANSFORMERS
240V cable input USA 115V Flat pin outlets

Ref.	£	P&P
75VA	£4.96	.90
75VA	£6.03	1.14
150VA	£8.48	1.14
250VA	£9.92	1.45
500VA	£15.73	1.64
750VA	£18.55	1.76
1000VA	£22.68	2.30
1500VA	£26.02	OA
2000VA	£37.65	OA

PANEL METERS

2 inch	4 inch
2' 0-50µA £5.50	4' 0-50µA £6.70
2' 0-100µA £5.50	4' 0-100µA £6.70
2' 0-500µA £5.50	4' 0-500µA £6.70
2' 0-1mA £5.50	4' 0-1mA £8.40
2' 50V £5.50	4' 0-50V £8.40

VU Indicator Panel 48.45 250µA a FSD **£2.60**
VU Indicator Edge 54x14 **£2.60**
Carnage 65p VAT 8%

7559 250-0-250V, 6 3V 12 9V-0-12 9V, **£6.50** & P.P. 90p

NEW RANGE TRANSFORMERS. Sec 45 36-0-36. 45 to give 36-0-36. 45-0-45V 72V or 90V.

2A	£9.89	PP £1.38	5A	£16.74	PP £2.15
3A	£11.47	PP £1.48	6A	£20.77	PP £2.30
4A	£13.90	PP £1.84			

PLUG-IN - SAVE BATTERIES
3300. 6 7 5. 9v at 300mA plugs direct into 13A socket (fused) **£3.30**
MEA 36, 3 4 5 6v at 100mA **£3.30**
A122 3V, 4 5V, 6V, 7.5V 9V 12V 500mA **£9.21**
P&P 55p. VAT 8%

ANTEX SOLDERING IRONS
15W **£3.75**, 25W **£3.95**
Stand for above **£1.40**, P&P 46p VAT 8%

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0A2	1.20	6B8G	0.75	6L8	0.60	12A7U	0.62	30PL13	1.30	CV6	0.60	HBL21	2.00
0B2	0.70	6BA6	0.65	6A19	2.00	12A7X	0.60	30PL14	1.50	CV8	1.00	EC32	1.00
0C3	0.50	6B8C	0.90	6D12	0.44	12B2G	0.44	30PL15	1.00	CV9	1.00	EC33	1.00
1A3	0.20	6BE6	0.70	6L10	0.20	12BA6	0.50	35A3	1.00	CY10	1.00	EC34	1.00
1A5GT	0.55	6BG6	1.00	6N7GT	0.70	12BE6	0.65	35C5	0.90	CY31	1.00	EC36	0.84
1A7GT	0.55	6BH6	1.10	6P12	1.00	12BF6	0.68	35D5	0.90	D1	0.50	EC38	0.84
1B3GT	0.55	6BU6	0.75	6P15	0.44	12BY7	1.15	35L6GT	0.90	D63	0.50	EC39	0.50
1C2	1.00	6B7A	0.85	6Q7G	0.75	12C7	0.55	35W4	0.55	DAC32	0.90	EC42	1.00
1D5	1.00	6BQ7A	1.00	6R7M	0.75	12D7	0.70	35Z5GT	0.70	DAP71	0.55	EC45	1.00
1E1	0.80	6BR7	1.00	6R7M	1.00	12K7G	0.50	50B5	0.85	DD4	0.90	EC46	1.00
1H4	0.25	6BR8	1.25	6SA7	0.70	12K8	0.75	50C5	0.70	DF33	0.75	EC47	1.00
1LD5	0.70	6BW6	3.75	6SC7GT	0.70	12M7G	0.50	50D6G	0.40	DF91	0.90	EC48	1.00
1LN5	0.70	6B7V	0.65	6S7	0.35	12S2GT	0.40	50E15	0.65	DF96	1.00	EC49	1.00
1N5GT	0.75	6BK6	0.40	6SH7	0.70	12S7	0.50	50L6GT	1.00	DH63	0.75	EC52	1.00
1R5	0.50	6BY7	0.45	6SJ7	0.70	12S7G	0.55	66K1U	1.00	DH76	0.50	EC54	1.00
1S4	0.50	6BZ6	1.50	6SK7	1.00	12SH7	0.50	72	0.70	DH77	0.60	EC55	1.00
1S5	0.35	6C4	0.50	6SK7GT	0.70	12S7	0.60	77	0.45	DH81	1.00	EC56	1.00
1T4	0.30	6C6	0.45	6S97	0.70	12SK7	0.60	85A2	1.40	DK32	0.60	EC58	0.72
1U5	0.70	6C7	0.60	6T2GT	0.90	12S2GT	0.40	85A3	1.40	DK40	1.00	EC59	0.55
1U6	0.85	6C10	1.00	6U7G	0.55	12S7	0.40	90C1	1.00	DK91	0.85	EC61	1.00
2G4	0.85	6C2A	0.65	6B	0.50	12S2GT	0.40	90C5	1.00	DK92	1.00	EC62	1.00
2K2	0.70	6C12	0.55	6V2G	0.50	12S7	0.75	108C4	1.40	DK96	1.00	EC63	1.00
3A4	0.55	6C6G	0.40	6X2	0.90	13D8	2.00	150C2	1.20	DL33	0.70	EC64	1.00
3B7	0.55	6C8A	0.90	6A4	0.95	14H7	0.75	21S8G	0.60	DL63	0.70	EC65	1.00
3D6	0.40	6C16	0.75	6X3GT	0.50	14S7	1.00	303	1.20	DL82	1.00	EC66	1.00
3K4	0.80	6C18	0.85	6Y7	0.65	14S7	1.00	303	1.20	DL83	1.00	EC67	1.00
3Q5GT	0.70	6CMT	1.00	6Y7G	1.25	18	1.25	807	1.10	DL94	1.00	EC68	1.00
3S4	0.65	6C56	0.75	7A7	1.00	19A05	0.65	956	0.50	DL96	1.00	EC69	1.00
3V4	1.00	6C65	0.90	7B6	1.00	19B2G6	1.00	1625	2.50	DM70	1.25	EC70	1.00
4C6	0.75	6D3	0.75	7B7	1.00	19C6	0.50	1821	1.00	DM71	1.75	EC71	1.00
4GK5	0.75	6DE7	0.90	7D6	2.00	19H1	4.00	5702	1.20	DWA/350	1.15	EC72	1.00
5C3B	0.75	6D7B	0.85	7E8	2.00	19H2	3.65	5783	1.20	DWA/350	1.15	EC73	1.00
5R4CY	1.00	6E4W	0.85	7H7	1.00	20D1	0.70	6657	2.00	DY97	0.52	EC74	1.00
5T4	2.00	6E5	1.00	7R7	2.00	20D4	2.50	6060	2.00	DY92	0.50	EC75	1.00
5U4G	1.00	6F1	0.90	7V7	2.00	20F2	0.85	6067	2.00	E00CC	4.75	EC76	1.00
5V4G	1.00	6F6G	0.70	7Y4	0.90	20L1	1.20	6146	4.70	E00CF	5.00	EC77	1.00
5Y3GT	0.65	6F12	0.70	7Z4	0.90	20P1	1.00	6463	2.00	E00E	5.00	EC78	1.00
5Z3	1.40	6F14	0.90	8B7	0.50	20P3	1.00	6566A	6.75	E10CC	2.00	EC79	1.00
5Z4	0.75	6F16	0.85	8B8	0.52	20P4	1.00	7025	2.00	E10CC	2.00	EC80	1.00
5Z4GT	1.00	6F16	1.00	8B8W	0.50	20P5	1.50	7193	0.90	E10E	2.00	EC81	1.00
6/30L2	0.90	6F18	0.60	9D7	0.70	25A6G	1.00	7475	1.20	E88CC	1.20	EC82	1.00
6A6C	1.40	6F23	1.00	9U8	0.45	25L6G	1.00	9002	0.55	E92CC	1.50	EC83	1.00
6AC7	0.70	6F24	1.00	10C2	0.70	25Y3	0.90	9006	0.65	E10CC	5.00	EC84	1.00
6A75	0.35	6F25	1.00	10C14	0.60	25Z4G	0.50	9183A	1.50	E10E	5.00	EC85	1.00
6A7G	0.70	6F26	0.85	7E8	2.00	19H2	3.65	5783	1.20	E10E	5.00	EC86	1.00
6A7H	0.70	6F28	0.85	10DE7	0.80	25Z6G	0.60	AC2PEN	1.00	E18CC	4.60	EC87	1.00
6A7J	0.60	6F32	1.00	10F1	1.00	28D7	2.00	AC2PENDD	1.00	E280F	12.50	EC88	1.00
6A7K	0.55	6G6G	1.00	10F9	0.65	30A5	0.75	1.00	1.00	E280F	12.50	EC89	1.00
6A7L	0.45	6G18A	0.80	10F18	0.65	30C1	0.90	AC6/PEN	1.00	E1148	0.90	EC90	1.00
6A7M	1.50	6G35	0.75	10F1D1	0.39	30C15	1.00	AC/P	1.50	E1148	0.90	EC91	1.00
6A7N	0.48	6A7N	0.48	10F14	0.50	30C19	1.00	AC/PENGT	1.00	E1148	0.90	EC92	1.00
6A7O	1.25	6G17	0.90	10D11	0.75	30C18	2.25	1.00	1.00	E1148	0.90	EC93	1.00
6A7P	0.60	6G18	0.50	10D12	0.45	30F5	0.70	AC/THI	1.00	E1148	0.90	EC94	1.00
6AM7A	0.70	6U5GT	0.65	10P12	0.75	30L1	0.39	AL60	1.50	E1148	0.90	EC95	1.00
6AN8	0.78	6G	0.35	10P13	0.60	30L15	0.75	AR/P	1.00	E1148	0.90	EC96	1.00
6A7Q	0.75	6G7	0.50	10P14	2.50	30P1	0.90	AT/P	0.50	E1148	0.90	EC97	1.00
6A7R	0.50	6A7R	0.50	10P15	0.90	30P4M	0.90	AT/P	0.50	E1148	0.90	EC98	1.00
6A7S	0.65	6I18A	0.90	12A6	1.00	30P12	0.74	AZ31	1.00	E1148	0.90	EC99	1.00
6A7T	1.50	6K7G	0.50	12A6C	0.90	30P19	0.90	AZ41	0.50	E1148	0.90	EC100	1.00
6A7U	0.60	6K8G	0.50	12A6D	0.90	30P4	0.90	B36	2.00	E1148	0.90	EC101	1.00
6A7V	0.62	6K8GT	0.55	12A6E	0.90	30P16	0.50	B719	0.50	E1148	0.90	EC102	1.00
6A7W	0.65	6L1	2.50	12A7E	0.45	30P17	0.50	B729	0.90	E1148	0.90	EC103	1.00
6A7X	1.15	6L2	1.50	12A7G	0.50	30P18	0.50	B733	0.30	E1148	0.90	EC104	1.00
6A7Y	0.75	6L12	0.50	12A7H	0.60	30P12	0.62	CL33	2.00	E1148	0.90	EC105	1.00

EL81	1.00	N308	0.96	PM84	0.75	UY41	0.70	XY21	0.35	BY212	0.30	OC38	0.50
EL83	1.00	N339	1.25	PY31	0.50	UY42	0.70	XY22	0.35	BY213	0.30	OC41	0.58
EL84	0.65	N379	0.50	PY32/2	0.50	UY85	0.70	XY25	2.00	AD140	0.50	OC42	0.73
EL86	0.60	N709	0.48	PY80	0.50	U10	1.00	XY26	2.00	AD161	0.53	OC43	1.37
EL87	0.90	P81	0.90	PY81	0.60	U11/214	1.15	XY27	0.75	AF102	1.04	OC44	0.12
EL95	0.75	PAB30	0.45	PY82	0.40	U11	1.00	XY28	0.90	AF125	0.58	OC65	0.31
EL360	2.50	PC86	0.90	PY83	0.60	U17	1.00	XY29	1.00	AF170	0.30	OC70	0.14
EL506	2.00	PC88	0.90	PY88	1.12	U18/20	2.50	XY30	1.00	AF175	0.30	OC71	0.20
EL509	2.50	PC92	0.65	PY901	0.50	U19	4.00	XY31	0.55	AF177	0.23	OC72	0.13
EM40	1.00	PC95	1.00	PY902	2.05	U22	0.85	XY32	1.00	AF178	0.35	OC73	0.26
EM81	1.00	PC97	0.75	PY90A	2.05	U25	1.00	XY33	1.00	AF179	0.36	OC74	0.13
EM82	1.00	PC98	0.65	PY90B	2.05	U28	0.90	XY34	0.70	AF179	0.36	OC76	0.18
EM84	1.00	PC99	0.39	PY901	0.60	U31	0.50	XY35	0.70	AF179	0.36	OC77	0.12
EM85	1.20	PC99	0.47	PZ30	0.50	U33	1.75	XY36	0.70	AF178	0.29	OC78	0.18
EM87	1.45	PC98	0.61	Q2P1	1.10	U35	1.75	XY37	1.00	AF180	0.56	OC79	0.10
EM88	0.89	Q0V03/10	0.37	2.00	0.43	0.90	0.90	XY38	1.00	AF186	0.64	OC81	0.20
EM89	0.90	PC98	0.60	Q575	2.00	1.00	1.00	XY39	0.50	AF229	0.44	OC82	0.13
EM90	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY40	0.50	AF229	0.44	OC83	0.13
EM91	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY41	0.50	AF229	0.44	OC84	0.13
EM92	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY42	0.50	AF229	0.44	OC85	0.13
EM93	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY43	0.50	AF229	0.44	OC86	0.13
EM94	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY44	0.50	AF229	0.44	OC87	0.13
EM95	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY45	0.50	AF229	0.44	OC88	0.13
EM96	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY46	0.50	AF229	0.44	OC89	0.13
EM97	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY47	0.50	AF229	0.44	OC90	0.13
EM98	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY48	0.50	AF229	0.44	OC91	0.13
EM99	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY49	0.50	AF229	0.44	OC92	0.13
EM100	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY50	0.50	AF229	0.44	OC93	0.13
EM101	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY51	0.50	AF229	0.44	OC94	0.13
EM102	1.00	PC98	0.60	Q575	2.00	1.00	1.00	XY52	0.50	AF229	0.44	OC95	0.13
EM103	1.00	PC98	0.6										

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EQUIPMENT AND SERVICES**

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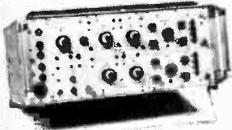
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Calibrated & Guaranteed 12 Months

GOULD ADVANCE

Double Pulse Generator PG.56. Pulse Amplitude 0.1V-10V Sq. wave 0-10V. Rise Time 10nsec (typically) **£100.00**



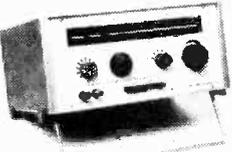
Pulse Generator PG59 (Type CT600). Produces a pulse controllable in frequency, width, amplitude, delay and rise and fall times. Also produces prepulse and gate outputs. Repetition frequency 1Hz to 10MHz. Pulse width 25nsec to 1sec. Rise and fall times 12nsec to 1 sec. Delay: 25usec to 1sec. Trigger: Internal, external and single shot **£595.00**

LR100X-Y Plotter. Specification, Main frame, Frequency response 1Hz. Plug-in amplifiers, LR120 general purpose amplifier, LR121 time base. Full specification on request **£235.00**

AF Signal Generator H1E. Specification, Frequency range, 15Hz-50KHz AF Output: Sine Wave 200µV to 20V r.m.s. ±2 dB. Square Wave 1.4mV to 14V peak to peak approximately **£80.00**

Signal Generator J3. Frequency range 10Hz to 100kHz Sine & Square output. Metered o/p **£150.00**

Signal Generator J4. Frequency range 10Hz to 100kHz Sine and square o/p **£140.00**



Wide range oscillator SG 67A. Frequency range 1Hz to 1MHz. Sine and square wave. **£95.00**

Low Distortion Oscillator SG 68A. 1.5Hz-150kHz. Output waveforms: Low distortion - <0.01% (40Hz-20kHz). Low distortion - <0.05% (20kHz-150kHz). Square wave. Typically 600Ω impedance. Max. output 4V (r.m.s.). Battery operated. **£175.00**

True R.M.S. Voltmeter DRM6. 100mV-1kV (6 ranges) 100% over-range. C.M.R. 90dB (DC-50Hz) **£395.00**

Frequency Counter TC16. 5Hz-80MHz 5 digit display, 1MHz crystal. Frequency and count functions **£110.00**

Timer Counter TC17A. 6 digit, DC-80MHz. Gate times 10µs to 10s in decade steps. Sensitivity 25mV (r.m.s.) sine wave. Overload protected **£215.00**

Timer Counter TC18. Specification, Frequency Measurement, Input 1 10Hz to 100MHz. Input 2: 10MHz-512MHz **£245.00**

Timer Counter TC22. Measures Frequency DC-100MHz 6 digit, time period, period average, count, totalise, pulse width, ratio **£275.00**

Digital Multimeter DMM3. 3½ digit Maximum reading 1999. 25 ranges of measurement, AC and DC volts, AC and DC current, Resistance, Dual Polarity, Remote hold **£115.00**

STRIP CHART RECORDER OMNISCRIIBE A.5000 Series. One & Two Pen. "As New Condition." Full specification on request **£200.00**

A 5110-1 **£200.00**

- A 5112-2 **£225.00**
- A 5123-1 **£300.00**
- A 5120-5 **£300.00**
- A 5132-4 **£320.00**
- A 5221-4 **£400.00**
- A 5212-4 **£380.00**
- A 5230-4 **£450.00**
- A 5222-2 **£410.00**

Off Air Frequency Standard OFS-1 **£120.00**

VHF Square Wave Generator SG21. 10KHz-100MHz **£60.00**

A.F. Signal Generator J2E. 15Hz-50kHz o/p (600Ω) 0.1mW-1W **£120.00**

A.C. Millivoltmeter VM77C. 0.001V-300V f.s.d. (12 ranges) 15Hz-4.5MHz **£55.00**

Timer Counter TC14. 9 digit Display storage, DC-250MHz. Time gate selectable 0.1µs-100s. Multiple period average 10-100. Sensitivity 10mV, 100mV, 500mV. Overload protected **£400.00**

Timer Counter TC15. 9 digit with storage and plug-in capability, DC-250MHz. Spec. similar to TC14 **£500.00**

Plug In Unit TC15 P1. 1MHz-500MHz 10mV-1V. Full 500MHz display with 1Hz resolution in only 2 secs **£150.00**

PG52B Modular Pulse Generator. Complete with full compliment of modules. Prestige condition **£700.00**

TEKTRONIX

Bench Oscilloscope Type 531A c/w dual trace vertical plug-in unit CA. DC-13.5MHz. Sensitivity 50mV-20V/Div. Time base ranges 100ns to 5S/Div. Timebase modes, A, X5. Internal voltage calibrator 0.2mV-100V 1kHz square wave **£290.00**

Bench Oscilloscope Type 647A c/w dual trace vertical plug in unit 10A2A & delayed time base plug in unit 11B2A. DC-100MHz. Sensitivity 10mV-20V/Div. Time base ranges 100ns-5s/Div on A & B. Time base modes A only, Intensified, delayed sweep, X10. Internal voltage calibrator 0.2mV-100V 1kHz square wave **£1200.00**

Bench Oscilloscope Type 585A c/w dual trace vertical plug in unit Type 82 DC-80MHz. Sensitivity 10mV-50V/Div. Time base ranges A 50ns-2s/Div, 2µs-1s/Div. Time base modes A, B, intensified, delayed, X5. Internal voltage calibrator 0.2mV-100V 1kHz square wave **£775.00**

Bench Oscilloscope Type 547 c/w dual trace vertical plug in unit 1A1. DC-50MHz. Sensitivity 5mV-20V/Div. Time base ranges, 100ns-5s/Div A and B. Time base modes A, B, intensified, delayed sweep, alternate sweep, X 2.5-10. Internal voltage calibrator 0.2mV-100V 1kHz square wave **£775.00**

Bench Oscilloscope 545B c/w dual trace vertical plug in unit CA. DC-24MHz, sensitivity 50mV to 20V/Div. Time base ranges A 100ns-5S/Div B2µs/1s/Div. Time base modes A, B, intensified, delayed sweep, X5. Internal voltage calibrator 0.2mV-100V 1kHz square waves **£555.00**

Bench Oscilloscope Type 543B c/w dual trace vertical plug in unit CA. As 545B & CA, without B time base. With X2-X100 Horizontal gain **£450.00**

Time Marker Generator 184. 16 marker intervals, 5 sinewave frequencies, 500MHz sinewave output. Crystal controlled oscillator, 10MHz ± 0.001% **£275.00**

2101 5ns Pulse Generator with Delay. 2.5Hz-25MHz repetition rate. Variable baseline offset. 5ns risetime and fall time. Paired, undelayed, delayed and output latched on modes. External gate input. Simultaneous positive and negative going pulses 10volts into 50Ω **£575.00**

2901 Time Mark Generator. 16 marker intervals, 4 sinewave frequencies, 500MHz sinewave output. Crystal controlled oscillator, 8 trigger pulse intervals **£450.00**

TV Waveform Monitor 525 **£165.00**

Plug In Unit Power Supply 132 **£120.00**

Spectrum Analyser Plug In 3L10 (for 560, and 564 series scopes), 1-36MHz **£415.00**

4-Trace Plug In Unit "M" (for use with 530, 540 & 580 series) **£250.00**

Differential Plug In Unit "W" **£325.00**

Oscilloscope Camera Unit C12 **£190.00**

Digital Readout Oscilloscope 567. Choice of plug-ins available. Mainframe price **£510.00**

Constant Amplitude Signal Generator 191. 350kHz-100MHz Sine wave 5mV-5.5V constant amplitude. 50kHz Amplitude reference **£350.00**

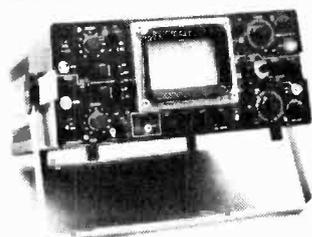
1L20 Spectrum Analyser. 10MHz-40GHz -70dBm to -110dBm sensitivity, dependent on frequency 40dB dynamic range. For use in 540 series main frame. **From £1050.00**

575 Semiconductor Curve Tracer. Displays dynamic characteristic curves and enables comparisons to be made - between two devices **£400.00**

7613 + 7L13 Spectrum Analyser with storage display unit, 1 KHz-1800 MHz 30 Hz-3MHz resolution, 70dB on screen dynamic range. -128dBm sensitivity, Auto Phase lock. Lin. 2dB and 10dB log display **P.O.A.**

TM504 + DC502 + TR502 Tracking Generator and Frequency Counter. Tracking Generator - 100KHz-180MHz 0 to -56dBm, o/p ± 0.5 dB flatness O/P Z 50 ohms Frequency Counter - 10Hz-550MHz 7 digit display. Sensitivity 100mV to 110 MHz into 1M ohm. 170mV to 550 MHz into 50 ohms. Designed for use with 7L13 analyser **P.O.A.**

DYNAMCO



Portable Solid State Scope 7500. DC-40MHz 10mV/div sensitivity X10 gain extends sensitivity to 1mV/div (3Hz-5MHz). Mixed & calibrated sweep delay. Dual trace. AC & DC coupled Z mod. BRAND NEW **£495.00**

Dual Trace Portable Oscilloscope Type D7210. DC-15MHz 10mV-5V/Div. Delayed Sweep Immaculate **£350.00**

Dual Trace Portable Oscilloscope Type D7200. Same as D7210 but with smaller CRT and no delay **£315.00**

WAYNE KERR

Universal Bridge B221A = CT530. 10S/948-8768 0.1% accuracy. Measures R +ve or -ve. G +ve or -ve. C (x1). L (1592Hz). Source frequency (1592Hz) (int). 50Hz-20Hz (Ext). Mains operated. Weight 11 kg **£275.00**

Low Impedance Adaptor Q221A. For use with B221A permitting impedance measurements of <10Ω to be made **£75.00**

Component Bridge B521 (Ministry Type CT375) A general purpose component bridge for measuring LC & R & combinations of R & Reactance. It can be used to measure components in situ. Capacitance 1pF to 5F Resistance 1 milliohm to 100Mohm. Inductance 1µH to 500kH **£115.00**

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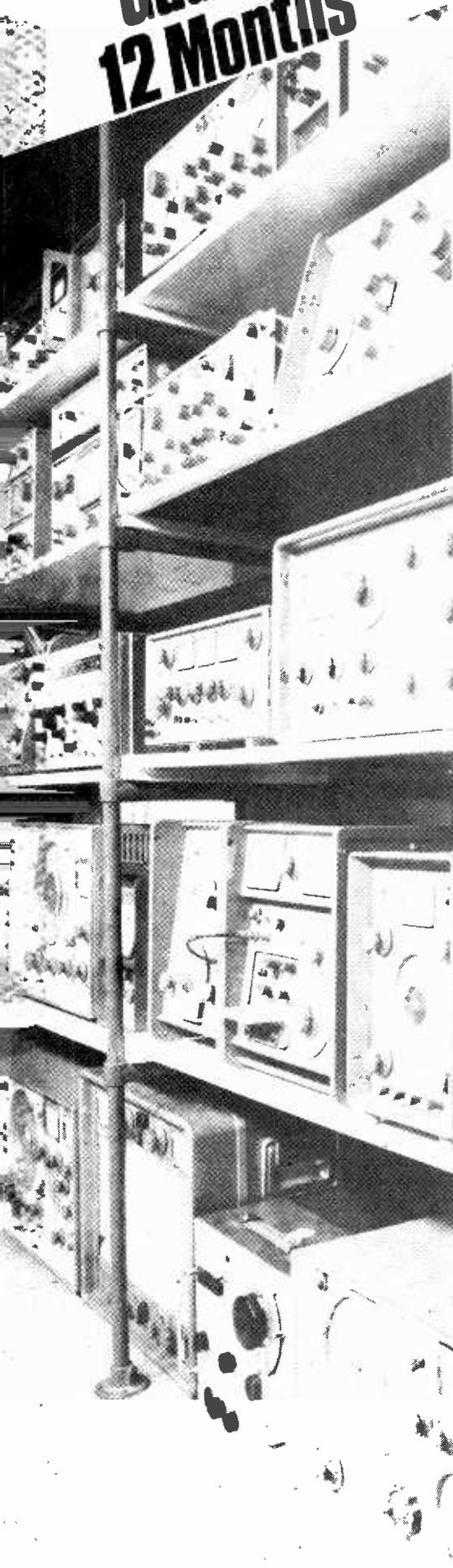
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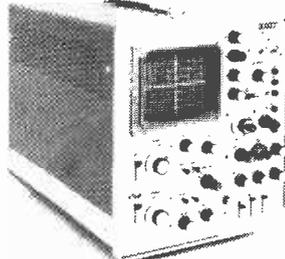
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COSSOR



Dual Trace Scope 4000. 50MHz 7nsec Rise Time 5mV cm sensitivity Calibrated sweep delay Gated trigger XY display 8 x 10cm display Unused **£495.00**

3100 Modular Oscilloscope. c/w 3121 3102 and 3122 modules DC 40 MHz 5mV/20V/div 0.2µ S-O 5S/div with X10 mag Full delayed sweep **£450.00**

TELEQUIPMENT

Dual Trace Moduler Oscilloscope D83/V4/S2A. DC 50 MHz 5m V 20V DIV Delayed Sweep Excellent condition **£610.00**

ROHDE & SCHWARZ

Stereocoder Type MSC. BN4192/2. Superb condition **£1,200.00**

Stereocoder MSDC. BN4193. Superb condition **£850.00**

Z-g Diagraph ZDU **£860.00**

Selective Level Voltmeter. USWV. 30,400 MHz 10µV 3V **£800.00**

U.H.F. Millivoltmeter. URV. 1 KHz-1.6 GHz **£235.00**

SWH M.F. Sweeper. 50 KHz-12MHz 50 V-2V o/p into 60 ohms 10-50-100 KHz markers 20 Hz sweep frequency Sweep width +0.05%-5% of range Excellent condition **£250.00**

HEWLETT PACKARD

Digital Multimeter 34702A. C/w 34740 display 4½ digit 4 ranges of AC & DC & 6 ranges of ohms A C = 45Hz-100kHz **£350.00**

U.H.F. Signal Generator 612A. 450-1200MHz in 1 band Output Voltage 0.1µV to 0.5V Output impedance 50Ω A M 0-90% Ext. A M Facility Pulse mod facility Various modern versions from **£900.00-£1250.00**

Frequency Doubler 10515A **£75.00**

Bench Oscilloscope 175A. C/w Plug Ins 1754A & 1781B 4 trace 40MHz band, width Sweep Delay **£550.00**

Programmable Step Attenuator 355E. DC-1000MHz 0-12 dB in 1dB steps Unused **£60.00**

Distortion Analyser 332A. 5Hz-600kHz Distortion levels 0.1%-100% are measured full scale in 7 ranges A M Detector operating from 550kHz-65MHz **£495.00**

DC Standard / Differential Voltmeter 740B. Designed for calibrating digital voltmeters, differential voltmeters, potentiometers, voltage dividers and for general standards lab application 6 digit resolution with discrete steps of 1 p p m of full scale. Full specification on request **£1,850.00**

A.C. Voltmeter 400F. 100V to 300V f s 14 ranges Freq range 20Hz-4MHz **£195.00**

Microwave Power Meter 430C. C/w Thermister mount 477B 10MHz-10GHz Full scale ranges 0.1, 0.3, 1, 3 & 10mW **£195.00**

Portable Oscilloscope 1700B. DC-35MHz 10mV/div sensitivity Dual Channel Solid State **£575.00**

S.H.F. Signal Generator 628A. 15-21GHz 50Ω **£495.00**

5½ Digit DMM Type 3490A DC/AC Volts Resistance Auto Ranging excellent condition **£650.00**

AC-DC Differential Voltmeter / DC Standard Type 741B **£1,095.00**

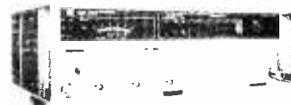
SHF Signal Generator Type 628A. 15-21 GHz **£495.00**

180A Plug in Oscilloscope. C/w 1802A and 1821A plug in units Dual trace DC-100MHz 5mV/10V/Div sensitivity Time base range 0.1µS-1S/Div with x 10 mag Full delayed sweep **£800.00**

202H. AM / FM Signal Generator. 54-216 MHz 0.1µV-0.2V 50 ohm o/p Z AM 0.50% FM 0-250 KHz **£645.00**

313A Distortion Analyser. Same specification as 332A but no AM detector **£450.00**

5300A/5306A Digital DMM / Counter. Counter 40Hz-10MHz 25-125mV sensitivity. Gate Time 1S and 0.1S. 6 digit display DMM AC-DC Volts 10-1000V 99999 F S Ohms 10Kohm-10Mohm **£425.00**



8640B opt 002 Am/FM Phase Locked Signal Generator with 6 digit frequency display. 500 KHz-1024MHz Stability 0.05 PPM / Hour +13 to -145 dBm o/p range o/p Z 50 ohms AM 0-100% FM 0.5-12 MHz Ext pulse mod AM and FM may be used simultaneously Counter may be used externally 1Hz-550MHz, 100mV i/p into 50 ohms Pristine condition **P.O.A.**

8690A R.F. Sweeper Main Frame with 8693/100 R.F. Unit 3-7.8 GHz 5mV o/p into 50 ohms Sweep time 0.015S-100S **£1,050.00**

FLUKE

Multifunction Counter 1900A-01. Same spec as 1900A but with battery pack **£215.00**

Multifunction Counter 1900A-02. Same spec as 1900A but with BCD o/p **£230.00**

Frequency Synthesiser 6011A. Performs function of an oscillator counter and level meter 10Hz 11MHz Output 0.4mV-5V r m s 7 digit LED display Accuracy ± parts in 10 for one year Freq coverage Full spec on request **£2650.00**

Frequency Synthesiser 6160A/DX. 4MHz 30MHz in 1Hz steps Output 1V into 50Ω Full spec on request **UNUSED BARGAIN PRICE £675.00**

Industrial Counter Totaliser 1941A. 5Hz-40MHz 40mV sensitivity R P M measurement **£150.00**

VHF, UHF Telecomm. Frequency Counter 1980A. C/w Battery Pack 5Hz-515MHz 6 digit LED Display 50mV sensitivity over whole range **£400.00**

8030A DMM. 3½ digit Scale length 1999 AC Volts 200mV-750V true RMS Frequency range 45Hz-1kHz. 100µV resolution DC Volts 200mV-1100V 100µV resolution. AC current 200µA-2A true RMS Frequency range 45Hz-5kHz. 100nA resolution DC Current 200µA-2A 100nA resolution Resistance 200 ohms-20Mohms 100m ohm resolution Battery operation **£130.00**

931B True RMS Differential Voltmeter. 10Hz-1MHz 0.01V-1100V rms Accuracy dependent on frequency I/P Z 1.0 Mohm **£1050.00**

883AB AC/DC Differential Voltmeter. 1mV-1100V 20Hz-1000Hz DC accuracy + (0.005% of I/P + 0.0002% of range + 5µV) AC accuracy + (0.1% of I/P + 25µV) 20Hz-5kHz **£975.00**

1900A Frequency Counter. 5Hz-80MHz 25mV sensitivity 6 digit display BRAND NEW **£175.00**

METRONIX (POLAND)

Function Generator G432. 1Hz-1 1MHz. Sine, Square & Triangle BRAND NEW. Full specification on request **£60.00**

RADIOMETER

Stereo Signal Generator SMG1C. Solid state, pushbutton operation. Full specification on request **£350.00**

AVO

Multimeter model 7 **£40.00**
Leads, prods & croc. clips for models 7, 8 & 9 **£4.50 set**

Multimeter Model 8 c/w lead Kit **£57.50**

Multimeter Model 7 c/w lead Kit **£44.50**

PHILIPS

Universal Counter Type PM 6615. 10Hz-1GHz 10mV rms, as new condition **£795.00**

Frequency Counter Type PM 6645B. 30Hz-512MHz 5mV rms, as new condition **£710.00**

Dual Trace Portable Oscilloscope Type PM 3260. DC-120MHz 5mV/2V/DIV Delayed Sweep as new condition **£1,385.00**

Dual Trace Portable Oscilloscope Type PM 3260E. As PM 3260 but no horizontal mag and prob. as new condition **£1,300.00**



Dual Trace Portable Oscilloscope Type PM 3240. DC-50 MHz 5mV/10V/DIV Delayed Sweep as new condition **£950.00**

4½ Digit DVM Type PM 2443. 10µV 1000V DC only High Accuracy Auto Ranging As new condition **£430.00**

4 Digit DMM Type PM 2424. AC DV Volt; and Current Resistance As new condition **£300.00**

Function Generator Type PM 5167 0.0001Hz 10MHz Various o/p waveforms 40V PK-PK into 50 ohm As new condition **£875.00**

Low Distortion LF Oscillator Type PN 5107. 0.02% As new condition **£210.00**

100MHz Single output Pulse Generator PM5775. Specification on request **£800.00**

100MHz Double Output Pulse Generator PM5776. Specification on request **£900.00**

50MHz Pulse Generator PM5712 **£495.00**

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Low Frequency Generator PM5105. 10Hz-100kHz sine & Square wave **£156.00**

Sine & Square Wave Oscillator PM5125. 10Hz-1MHz **£145.00**
PM5324 AM/FM Signal Generator. 100kHz-110MHz $5\mu V$ 50mV o/p into 75 ohms AM is 30% @ 1MHz FM is 25kHz @ 1kHz rate Very good condition **£450.00**

PM5326 AM/FM Signal Generator with Digital Frequency Display and Internal Sweep. 100 kHz-125MHz >math>5\mu V</math> 50mV into 75 ohms AM is 30% @ 1kHz FM is 22.5kHz @ 1kHz rate Sweep frequency 3-30 Hz As new condition **£695.00**



PM2513A D.M.M. AC-DC volts and current resistance 3 1/2 digit display As new condition **£95.00**

PM3010 Miniature oscilloscope. DC-5MHz 30mV-1V/div. 1μS-0.1S/div with X10 mag Battery operation Supplied with X10 probe and battery charger. As new condition **£335.00**

RACAL Universal Counter Timer 9838. Frequency, single and multiple period, time interval. Freq range 10Hz-100MHz **£285.00**

H.F. Communications Receiver RA117E. 1-30MHz Full specification on request **£350.00**

Modulation Meter 210A. 2.5-300MHz A.M range 0-100% F.M range 0 to ±100kHz **£245.00/£285.00**

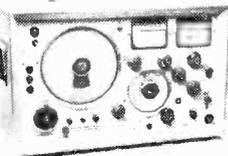
A.M./F.M. Modulation Meter 409. Freq. 3-1500MHz A.M. 0-100% F.M. 0-±600kHz **£345.00**

DUMONT Portable 'Scope 1100P. DC 100MHz at 5mV/cm Dual Trace 5ns/cm. Sweep rate Delayed time base Full 8 x 10cm display Calibrated X-Y to 5MHz **£825.00**

MARCONI INSTRUMENTS

F.M./A.M. Signal Generator TF995B/2. Frequency Range 200kHz to 220MHz in five bands Output 0.1μV-200mV **£675.00**

U.H.F. Signal Generator TF1060. Frequency range 450-1250MHz (1 band) Output 0.15μV to 445mV Output Impedance 50Ω Int. sine A.M.-1kHz Ext. pulse mod. **£400.00**



A.M. Signal Generator TF801D/1. Frequency Range 10MHz to 470MHz in five bands Output Attenuator 0.1μV to 1V Output Impedance 50Ω **£400.00-£750.00**

Carrier Deviation Meter TF791D. Carrier Freq range 4 to 1024MHz Deviation range up to ±1.25kHz Modulating Frequency range Up to 35kHz Late models **£295.00**

20MHz Sweep Generator TF1099. Video sweep output Lower limit 100kHz fixed. Upper limit continuously variable up to 20MHz. 0.3 to 3V p-p Z=75Ω Input & Output detector probes. Markers at 1MHz intervals **£295.00**

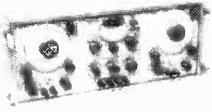
A.M. Signal Generator TF801D/8S. Same spec as TF801D/1S + freq counter o/p facility **£695.00**

Sensitive Valve Voltmeter TF2600. 12 ranges 1mV-300V f.s.d. 1% accuracy up to 500kHz Usable up to 10MHz **£175.00**

R.F. Electronic Voltmeter TF2604. 7 ranges 300mV-300V f.s.d. from 20Hz-1500MHz 8 ranges 300mV-1kV DC. 7 ranges Resistance 500Ω to 500MΩ **£225.00**

R-C Oscillator TF1101. Frequency Range 20Hz to 200kHz in four bands Output Attenuator 1mV to 20V Maximum Output 20V across external 600Ω load Output impedance 600Ω **£120.00**

Phase A.M. Signal Generator TF2003. 0.4-12MHz **£150.00**



Two-Tone Signal Source TF2005R. Frequency Range 20Hz-20kHz **£415.00**

Blanking and Sync Mixer TF2908. For 405, 525 and 625 line television systems Reshapes and mixes blanking and sync pulse waveform with video test waveform **£90.00**

L.F. Extension Unit TM6448. For use with Spectrum Analyser OA 1094A series. 100Hz to 3MHz **£200.00**

Wide Range R.C. Oscillator TF1370A. 10Hz to 10MHz sine wave. 10Hz to 100kHz square wave Output up to 31.6V **£275.00**

DC Multiplier TM5033A. HV probe up to 30kV Impedance 3000MΩ for use with TF1041 series or TF2604 **£25.00**

5 watt Dummy Load TM5582 for use with TF2604 **£25.00**

A.F. Oscillator TF2100. 20kc/s to 20Kc/s Extremely low distortion. Output Impedance 600Ω unbalanced. **£150.00**

M.F. Oscillator TF2101. 30c/s to 550kc/s Stable frequency Low distortion. Output Impedance 600Ω unbalanced. **£115.00**

F.M./A.M. Modulation meter TF2300S. Incorporating Oscillator TM8045/1 3.5 to 100MHz Full specification on request **£825.00**



M.F. Transmission Measuring Set TF2333. Frequency range 30Hz to 550kHz **£600.00**

F.M. Signal Generator TF1066B/6. 10-470MHz R.F. output 0.2μV-200mV in f Output Impedance 50Ω Modulation Internal AM 1kHz & 5kHz 0-40% External AM 30Hz to 15kHz 0-40% Internal FM 1kHz & 5kHz deviation 0 to 100kHz or up to 400kHz according to carrier freq range. External FM 30Hz to 100kHz deviation same as int. Crystal Calibrator facility **£685.00**

Variable Attenuator TF338C. 0.105dB Freqs up to 100kHz 600Ω impedance **£90.00**

M.F. Attenuator TF2162. DC-1MHz 0-11 dB in steps of 0.1 dB **£120.00**
Also TF1073A Spec as A/2S **£55.00**

10-Watt A.F. Power Meter TF893A. Frequency range 20Hz to 35kHz Five power ranges: 1mW to 10W full scale Impedance 2.5Ω to 20KΩ in 48 steps. Balanced or unbalanced inputs Direct calibration in watts and dBm **£165.00**

100-Watt R.F. Power Meter: TF1020A Series. Frequency range D.C. to 250 MHz. Two power ranges 50 and 100W full scale. Output Impedance 75Ω **£105.00**

In Situ Universal Bridge TF2701. Measures capacitance, capacitance with R in parallel Resistance & Inductance Int source 80Hz & 1KHz **£395.00**

R.F. Power Meter TF1020A/4M1. 75 ohms 1-50 watts & 2-100 watts DC-250MHz **£120.00**

R.F. Power Meter TF1020A/5M1. 50 ohms 1-50 watts 2-100 watts DC-250MHz **£135.00**

Output Test Set TF1065A. AF Power 10uW to 3W (5 ranges) Freq. Range 250Hz-10KHz RF Power 1-25W FM & AM measurements DC current & voltage measurements Full spec on request **£225.00**

FM Signal Generator Type TF 2006. 10-220 MHz FM up to 200 KHz deviation 0.2 μV-200 MV into 50 ohm **£950.00**

EDDYSTONE

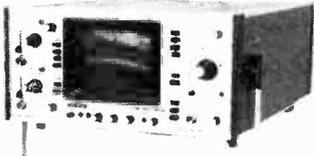
770U VHF Receiver. 180-390 MHz in 6 ranges AM/FM demod. Muting, AVC Noise limiting IF gain S meter **£265.00**

730/1A Communications Receiver. 480kHz-30 MHz in 5 ranges BFO noise limiting, AF filter, AVC, rf gain, S meter **£240.00**

880 Communications Receiver. 500kHz-30.5 MHz in 1 MHz wavebands BFO AGC, RF-IF gain, noise limiting, AF filter S meter **£325.00**

770R VHF Receiver. 19 MHz-165 MHz in 6 ranges AM/FM demod. Muting, noise limiting IF gain S meter **£275.00**

SOLARTRON



A100 Modular Oscilloscope c/w A101 and A112 modules Dual trace DC-30MHz 5mV/20V/div sensitivity Time base range 0.1μS-0.5S/div with x 5 mag Full delayed sweep **£380.00**

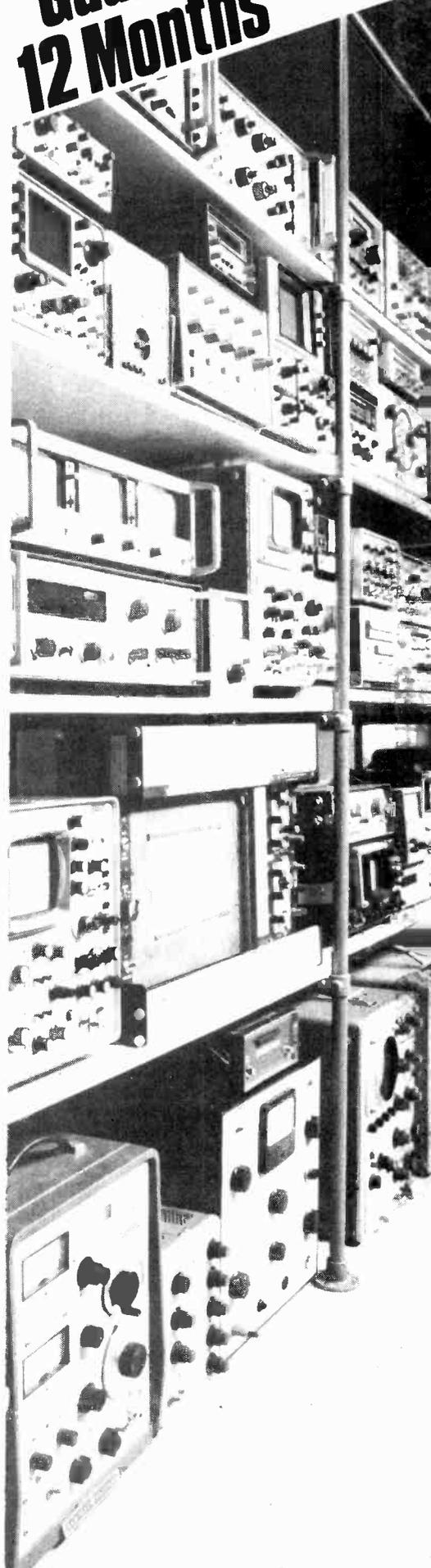
A100 Modular Oscilloscope c/w A101 and A111 Same spec as with A112 but no delayed sweep **£340.00**

CD1400 Modular Oscilloscope System. c/w CX1441 + CX1441 and C1442 Dual trace DC-15MHz 10mV-50V/div sensitivity Time base range 0.5μS-0.5S/div Available with various other modules **£180.00 to £205.00**

BRUEL & KJAER

2112 Audio Frequency Spectrometer. 22Hz-45KHz 1/3 and 1/1 octave filters A, B and C weighting RMS, average and peak measurement **£850.00**

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AA119 0.20	AS215 1.25	BC177 0.19	BD136 0.36*	BF336 0.50*	GJ3M 0.75	OA2207 0.65	OC205 1.75	ZTX504 0.20*	2N1671 1.50	2N3773 2.65
AA190 0.13	AS216 1.25	BC178 0.18	BD137 0.37*	BF337 0.38*	GJ3078A 1.50	OC16 1.65	OC206 1.75	ZTX531 0.20*	2N1893 0.33	2N3819 0.68
AA190 0.13	AS217 1.25	BC179 0.20	BD138 0.40*	BF338 0.40*	KS100A 0.40*	OC20 2.00	OC207 1.25	ZTX550 0.16*	2N2147 1.40	2N3820 0.66
AA190 0.13	AS218 1.25	BC180 0.11*	BD140 0.47*	BF339 0.57*	MJE340 0.58	OC22 2.50	OC208 2.50	OC209 1.25	2N2148 1.45	2N3823 0.40*
AA213 0.25	AS221 1.50	BC182 0.11*	BD141 2.00	BF340 1.38	MJE370 0.65	OC23 2.75	OC210 2.75	OC211 2.25	2N2218 0.33	2N3866 1.00
AA215 0.25	AU110 1.70*	BC183 0.11*	BD142 2.00	BF341 0.25*	MJE371 0.61	OC24 3.50	R2008B 2.25*	1N4001 0.07	2N2219 0.42	2N3904 0.21*
AC107 0.75	BA145 1.15*	BC212 0.14*	BD182 1.48	BF510 1.90	MJE520 0.65	OC25 0.90	R2009 2.25*	1N4002 0.07	2N2220 0.33	2N3905 0.22*
AC125 0.30	BA149 1.15*	BC214 0.17*	BD237 0.80	BF511 0.90	MJE521 0.75	OC26 0.90	R2010B 2.25*	1N4003 0.08	2N2221 0.22	2N3906 0.22*
AC126 0.25	BA154 0.10	BC217 0.17*	BD238 0.85	BF512 0.26	MJE525 1.25	OC28 2.00	T1C44 0.38	1N4004 0.09	2N2222 0.25	2N4058 0.15*
AC127 0.25	BA155 0.12	BC219 0.17*	BD239 0.85	BF513 0.26	MJE525 1.25	OC29 2.00	T1C22D 1.30	1N4005 0.13	2N2223 0.17	2N4059 0.15*
AC128 0.25	BA156 0.13	BC220 0.17*	BD240 0.85	BF514 0.25	MJE526 0.38	OC30 2.50	T1L208 0.25	1N4006 0.15	2N2268 0.25	2N4060 0.20*
AC141 0.20	BA162 0.05	BC301 0.65	BD241 1.42	BF515 0.25	MJE526 0.38	OC31 0.50	T1P29A 0.50	1N4007 0.17	2N2269A 0.21	2N4061 0.17*
AC142 0.20	BA163 0.07	BC302 0.60	BD242 1.42	BF516 0.25	MJE526 0.38	OC32 0.50	T1P30A 0.60*	1N4008 0.15	2N2270 0.21	2N4062 0.17*
AC142K 0.35	BA163 0.07	BC302 0.60	BD243 1.48	BF517 0.26	MJE526 0.38	OC33 1.50	T1P31A 0.42	1N4009 0.17	2N2271 0.22	2N4063 0.22*
AC176 0.20	BC108 0.12	BC303 0.60	BD244 1.48	BF518 0.26	MJE526 0.38	OC34 1.50	T1P32A 0.75	1N4010 0.16	2N2272 0.22	2N4064 0.22*
AC187 0.25	BC109 0.13	BC304 0.60	BD245 1.48	BF519 0.26	MJE526 0.38	OC35 1.50	T1P33A 1.00	1N4011 0.16	2N2273 0.22	2N4065 0.22*
AC188 0.25	BC113 0.15*	BC305 0.60	BD246 1.48	BF520 0.26	MJE526 0.38	OC36 1.50	T1P34A 1.20	1N4012 0.16	2N2274 0.22	2N4066 0.22*
AC187 0.25	BC114 0.15*	BC306 0.60	BD247 1.48	BF521 0.26	MJE526 0.38	OC37 0.50	T1P41A 0.70	1N4013 0.16	2N2275 0.22	2N4067 0.22*
AC187 0.25	BC115 0.15*	BC307 0.60	BD248 1.48	BF522 0.26	MJE526 0.38	OC38 0.50	T1P42A 0.90	1N4014 0.16	2N2276 0.22	2N4068 0.22*
AC187 0.25	BC116 0.15*	BC308 0.60	BD249 1.48	BF523 0.26	MJE526 0.38	OC39 0.50	T1P43A 1.20	1N4015 0.16	2N2277 0.22	2N4069 0.22*
AC187 0.25	BC117 0.15*	BC309 0.60	BD250 1.48	BF524 0.26	MJE526 0.38	OC40 0.50	T1P44A 1.20	1N4016 0.16	2N2278 0.22	2N4070 0.22*
AC187 0.25	BC118 0.15*	BC310 0.60	BD251 1.48	BF525 0.26	MJE526 0.38	OC41 0.50	T1P45A 1.20	1N4017 0.16	2N2279 0.22	2N4071 0.22*
AC187 0.25	BC119 0.15*	BC311 0.60	BD252 1.48	BF526 0.26	MJE526 0.38	OC42 0.50	T1P46A 1.20	1N4018 0.16	2N2280 0.22	2N4072 0.22*
AC187 0.25	BC120 0.15*	BC312 0.60	BD253 1.48	BF527 0.26	MJE526 0.38	OC43 0.50	T1P47A 1.20	1N4019 0.16	2N2281 0.22	2N4073 0.22*
AC187 0.25	BC121 0.15*	BC313 0.60	BD254 1.48	BF528 0.26	MJE526 0.38	OC44 0.50	T1P48A 1.20	1N4020 0.16	2N2282 0.22	2N4074 0.22*
AC187 0.25	BC122 0.15*	BC314 0.60	BD255 1.48	BF529 0.26	MJE526 0.38	OC45 0.50	T1P49A 1.20	1N4021 0.16	2N2283 0.22	2N4075 0.22*
AC187 0.25	BC123 0.15*	BC315 0.60	BD256 1.48	BF530 0.26	MJE526 0.38	OC46 0.50	T1P50A 1.20	1N4022 0.16	2N2284 0.22	2N4076 0.22*
AC187 0.25	BC124 0.15*	BC316 0.60	BD257 1.48	BF531 0.26	MJE526 0.38	OC47 0.50	T1P51A 1.20	1N4023 0.16	2N2285 0.22	2N4077 0.22*
AC187 0.25	BC125 0.15*	BC317 0.60	BD258 1.48	BF532 0.26	MJE526 0.38	OC48 0.50	T1P52A 1.20	1N4024 0.16	2N2286 0.22	2N4078 0.22*
AC187 0.25	BC126 0.15*	BC318 0.60	BD259 1.48	BF533 0.26	MJE526 0.38	OC49 0.50	T1P53A 1.20	1N4025 0.16	2N2287 0.22	2N4079 0.22*
AC187 0.25	BC127 0.15*	BC319 0.60	BD260 1.48	BF534 0.26	MJE526 0.38	OC50 0.50	T1P54A 1.20	1N4026 0.16	2N2288 0.22	2N4080 0.22*
AC187 0.25	BC128 0.15*	BC320 0.60	BD261 1.48	BF535 0.26	MJE526 0.38	OC51 0.50	T1P55A 1.20	1N4027 0.16	2N2289 0.22	2N4081 0.22*
AC187 0.25	BC129 0.15*	BC321 0.60	BD262 1.48	BF536 0.26	MJE526 0.38	OC52 0.50	T1P56A 1.20	1N4028 0.16	2N2290 0.22	2N4082 0.22*
AC187 0.25	BC130 0.15*	BC322 0.60	BD263 1.48	BF537 0.26	MJE526 0.38	OC53 0.50	T1P57A 1.20	1N4029 0.16	2N2291 0.22	2N4083 0.22*
AC187 0.25	BC131 0.15*	BC323 0.60	BD264 1.48	BF538 0.26	MJE526 0.38	OC54 0.50	T1P58A 1.20	1N4030 0.16	2N2292 0.22	2N4084 0.22*
AC187 0.25	BC132 0.15*	BC324 0.60	BD265 1.48	BF539 0.26	MJE526 0.38	OC55 0.50	T1P59A 1.20	1N4031 0.16	2N2293 0.22	2N4085 0.22*
AC187 0.25	BC133 0.15*	BC325 0.60	BD266 1.48	BF540 0.26	MJE526 0.38	OC56 0.50	T1P60A 1.20	1N4032 0.16	2N2294 0.22	2N4086 0.22*
AC187 0.25	BC134 0.15*	BC326 0.60	BD267 1.48	BF541 0.26	MJE526 0.38	OC57 0.50	T1P61A 1.20	1N4033 0.16	2N2295 0.22	2N4087 0.22*
AC187 0.25	BC135 0.15*	BC327 0.60	BD268 1.48	BF542 0.26	MJE526 0.38	OC58 0.50	T1P62A 1.20	1N4034 0.16	2N2296 0.22	2N4088 0.22*
AC187 0.25	BC136 0.15*	BC328 0.60	BD269 1.48	BF543 0.26	MJE526 0.38	OC59 0.50	T1P63A 1.20	1N4035 0.16	2N2297 0.22	2N4089 0.22*
AC187 0.25	BC137 0.15*	BC329 0.60	BD270 1.48	BF544 0.26	MJE526 0.38	OC60 0.50	T1P64A 1.20	1N4036 0.16	2N2298 0.22	2N4090 0.22*
AC187 0.25	BC138 0.15*	BC330 0.60	BD271 1.48	BF545 0.26	MJE526 0.38	OC61 0.50	T1P65A 1.20	1N4037 0.16	2N2299 0.22	2N4091 0.22*
AC187 0.25	BC139 0.15*	BC331 0.60	BD272 1.48	BF546 0.26	MJE526 0.38	OC62 0.50	T1P66A 1.20	1N4038 0.16	2N2300 0.22	2N4092 0.22*
AC187 0.25	BC140 0.15*	BC332 0.60	BD273 1.48	BF547 0.26	MJE526 0.38	OC63 0.50	T1P67A 1.20	1N4039 0.16	2N2301 0.22	2N4093 0.22*
AC187 0.25	BC141 0.15*	BC333 0.60	BD274 1.48	BF548 0.26	MJE526 0.38	OC64 0.50	T1P68A 1.20	1N4040 0.16	2N2302 0.22	2N4094 0.22*
AC187 0.25	BC142 0.15*	BC334 0.60	BD275 1.48	BF549 0.26	MJE526 0.38	OC65 0.50	T1P69A 1.20	1N4041 0.16	2N2303 0.22	2N4095 0.22*
AC187 0.25	BC143 0.15*	BC335 0.60	BD276 1.48	BF550 0.26	MJE526 0.38	OC66 0.50	T1P70A 1.20	1N4042 0.16	2N2304 0.22	2N4096 0.22*
AC187 0.25	BC144 0.15*	BC336 0.60	BD277 1.48	BF551 0.26	MJE526 0.38	OC67 0.50	T1P71A 1.20	1N4043 0.16	2N2305 0.22	2N4097 0.22*
AC187 0.25	BC145 0.15*	BC337 0.60	BD278 1.48	BF552 0.26	MJE526 0.38	OC68 0.50	T1P72A 1.20	1N4044 0.16	2N2306 0.22	2N4098 0.22*
AC187 0.25	BC146 0.15*	BC338 0.60	BD279 1.48	BF553 0.26	MJE526 0.38	OC69 0.50	T1P73A 1.20	1N4045 0.16	2N2307 0.22	2N4099 0.22*
AC187 0.25	BC147 0.15*	BC339 0.60	BD280 1.48	BF554 0.26	MJE526 0.38	OC70 0.50	T1P74A 1.20	1N4046 0.16	2N2308 0.22	2N4100 0.22*
AC187 0.25	BC148 0.15*	BC340 0.60	BD281 1.48	BF555 0.26	MJE526 0.38	OC71 0.50	T1P75A 1.20	1N4047 0.16	2N2309 0.22	2N4101 0.22*
AC187 0.25	BC149 0.15*	BC341 0.60	BD282 1.48	BF556 0.26	MJE526 0.38	OC72 0.50	T1P76A 1.20	1N4048 0.16	2N2310 0.22	2N4102 0.22*
AC187 0.25	BC150 0.15*	BC342 0.60	BD283 1.48	BF557 0.26	MJE526 0.38	OC73 0.50	T1P77A 1.20	1N4049 0.16	2N2311 0.22	2N4103 0.22*
AC187 0.25	BC151 0.15*	BC343 0.60	BD284 1.48	BF558 0.26	MJE526 0.38	OC74 0.50	T1P78A 1.20	1N4050 0.16	2N2312 0.22	2N4104 0.22*
AC187 0.25	BC152 0.15*	BC344 0.60	BD285 1.48	BF559 0.26	MJE526 0.38	OC75 0.50	T1P79A 1.20	1N4051 0.16	2N2313 0.22	2N4105 0.22*
AC187 0.25	BC153 0.15*	BC345 0.60	BD286 1.48	BF560 0.26	MJE526 0.38	OC76 0.50	T1P80A 1.20	1N4052 0.16	2N2314 0.22	2N4106 0.22*
AC187 0.25	BC154 0.15*	BC346 0.60	BD287 1.48	BF561 0.26	MJE526 0.38	OC77 0.50	T1P81A 1.20	1N4053 0.16	2N2315 0.22	2N4107 0.22*
AC187 0.25	BC155 0.15*	BC347 0.60	BD288 1.48	BF562 0.26	MJE526 0.38	OC78 0.50	T1P82A 1.20	1N4054 0.16	2N2316 0.22	2N4108 0.22*
AC187 0.25	BC156 0.15*	BC348 0.60	BD289 1.48	BF563 0.26	MJE526 0.38	OC79 0.50	T1P83A 1.20	1N4055 0.16	2N2317 0.22	2N4109 0.22*
AC187 0.25	BC157 0.15*	BC349 0.60	BD290 1.48	BF564 0.26	MJE526 0.38	OC80 0.50	T1P84A 1.20	1N4056 0.16	2N2318 0.22	2N4110 0.22*
AC187 0.25	BC158 0.15*	BC350 0.60	BD291 1.48	BF565 0.26	MJE526 0.38	OC81 0.50	T1P85A 1.20	1N4057 0.16	2N2319 0.22	2N4111 0.22*
AC187 0.25	BC159 0.15*	BC351 0.60	BD292 1.48	BF566 0.26	MJE526 0.38	OC82 0.50	T1P86A 1.20	1N4058 0.16	2N2320 0.22	2N4112 0.22*
AC187 0.25	BC160 0.15*	BC352 0.60	BD293 1.48	BF567 0.26	MJE526 0.38	OC83 0.50	T1P87A 1.20	1N4059 0.16	2N2321 0.22	2N4113 0.22*
AC187 0.25	BC161 0.15*	BC353 0.60	BD294 1.48	BF568 0.26	MJE526 0.38	OC84 0.50	T1P88A 1.20	1N4060 0.16	2N2322 0.22	2N4114 0.22*
AC187 0.25	BC162 0.15*	BC354 0.60	BD295 1.48	BF569 0.26	MJE526 0.38	OC85 0.50	T1P89A 1.20	1N4061 0.16	2N2323 0.22	2N4115 0.22*
AC187 0.25	BC163 0.15*	BC355 0.60	BD296 1.48	BF570 0.26	MJE526 0.38	OC86 0.50	T1P90A 1.20	1N4062 0.16	2N2324 0.22	2N4116 0.22*
AC187 0.25	BC164 0.15*	BC356 0.60	BD297 1.48	BF571 0.26	MJE526 0.3					

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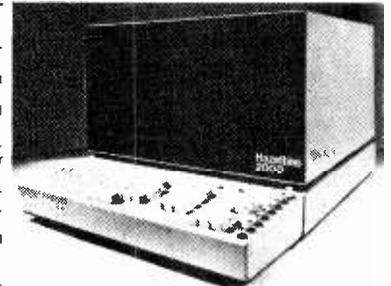
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SCREEN SIZE — 12" diagonal

SCREEN CAPACITY —
H1000 — 960 characters 80 per line x 12 lines
H1200 — 1920 characters 80 per line x 24 lines
CHARACTER GENERATION — 5 x 7 dot matrix 625 line raster
CHARACTER SET — 64 ASCII alphanumerics and symbols
CHARACTER SIZE —
H1000-1/8 inch (32cm) nominal height 3/32 inch (24cm) nominal width
H1200 — 1/10 inch (25cm) nominal height 3/32 inch (24cm) nominal width
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REFRESH RATE — 50 fields per second
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MEMORY — High speed MOS refresh
STANDARD INTERFACE — CC ITT V-24 (EIA RS-232 B/C), 202C Optional

HAZELTINE H2000 SPECIFICATION

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CHARACTER GENERATION — 5 x 7 dot matrix 625 line raster
CHARACTER SET — 64 alphanumerics and symbols 32 ASCII control codes
KEYBOARD — Detachable, solid state, teletypewriter design; 10-key numeric cluster plus editing and cursor control keys
TRANSMISSION — Asynchronous. Switch-selectable for combinations of 5 standard rates, 75 to 9600 baud
OPERATING MODES — Switch-selectable, full duplex, half duplex or batch (buffered)
MEMORY TYPE — 2048 x 8 RAM
EDITING FEATURES — Full Cursor controls plus Insert/Delete Character, Insert/Delete Line, Clear Screen, Clear Foreground Data Only, Tab
STANDARD INTERFACE — CC ITT V-24 (EIA RS-232 B/C) or 202C Compatible
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48V AC	2 CO	8	85p
30V DC	2 CO	8	75p
24V DC	3 CO	11	85p
24V DC	2 CO	8	85p
24V DC	1 CO	8	85p
12V AC	3 CO	11	95p
12V AC	2 CO	8	85p
12V DC	3 CO	11	95p
10V DC	3 CO	11	95p

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COIL VOLTAGE	CONTACT	PRICE
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240V AC	2 CO	75p
240V AC	1 CO	65p
110V AC	3 CO	65p
110V AC	2 CO	65p
110V AC	1 CO	65p
50V AC	3 CO	65p
50V AC	2 CO	65p
50V AC	1 CO	65p
24V DC	2 CO	75p
12V DC	2 CO	75p
10V DC	3 CO	85p

PP 1pp per relay. Please add VAT 8% on total.

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8	250V	40p	25p
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4	350V	30p	15p
8	600V	35p	15p
8	500V	30p	15p
8	1500V	40p	35p
8	1000V	35p	15p
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2 4	360VAC	75p
2 5	360VAC	75p
2 7+0 1	700VAC	£1.25
3	440VAC	£1.00
3 5	250VAC	£1.00
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5	360VAC	£1.25
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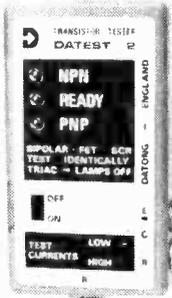
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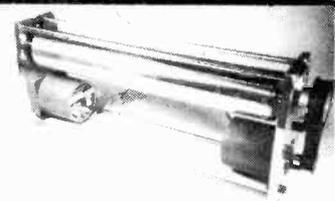
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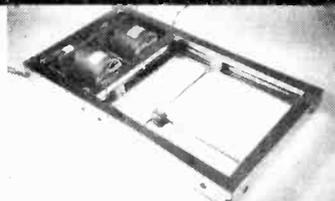
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X-Y PLOTTER ASSEMBLY

X-Y PLOTTER ASSEMBLY

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PAPER TAPE READER ASSEMBLY

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2N2905 0.31	2N3819 0.36	2N4870L 0.88	40488 0.82	BC136 0.21	BC307B 0.16
2N2906 0.25	2N3820 0.39	2N4871L 0.51	40440 0.70	BC137 0.22	BC308B 0.16
2N2907 0.25	2N3821 0.95	2N4898 1.55	40512 1.70	BC138 0.44	BC309C 0.16
2N2923 0.17	2N3827 0.27	2N4901 1.85	40594 0.87	BC140 0.30	BC327 0.22
2N2924 0.17	2N3854A 0.30	2N4902 2.20	40595 0.98	BC141 0.32	BC328 0.20
2N2925 0.19	2N3855 0.30	2N4903 2.75	40573 0.80	BC142 0.32	BC337 0.20
2N3011 0.37	2N3856A 0.19	2N4904 1.85	AC126 0.48	BC147 0.13	BC414 0.17
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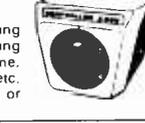
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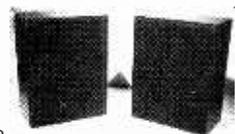
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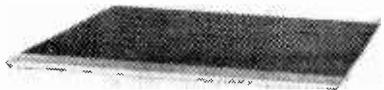
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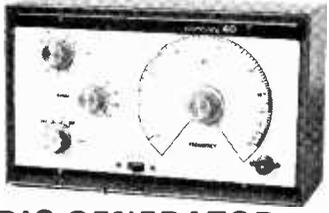
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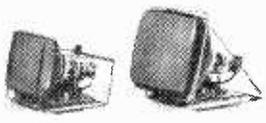
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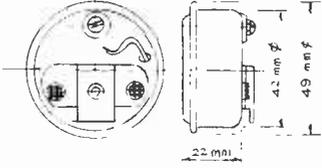
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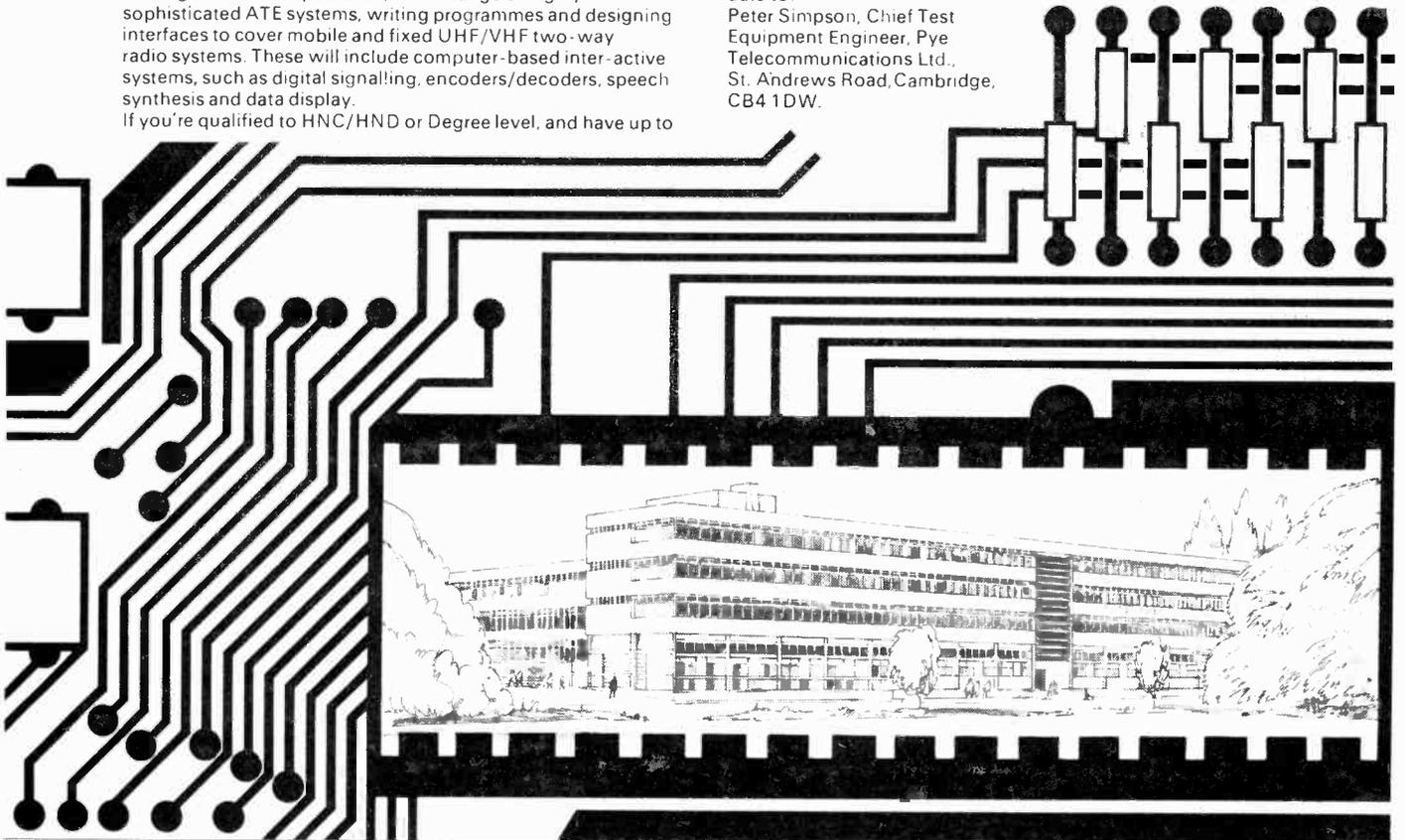
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R & D Engineers at senior and intermediate levels

required to work on digital and cable television systems for the domestic and surveillance market.

Engineers should hold a degree, HNC or equivalent qualification and have some knowledge of either HF video or digital circuit design.

Salaries will be commensurate with qualifications, age and experience.

Fringe benefits include a contributory pension, life assurance scheme, subsidised canteen, etc.

If you are seeking an enjoyable position in R & D, write, giving full details of your career to date, or telephone Dr. G. O. Towler, B.Sc., Ph.D. (Manager), Research and Development Establishment, British Relay Ltd., Cleeve Road, Leatherhead, Surrey, Tel. 76056

8469



SULTANATE OF OMAN

COLOUR TELEVISION SERVICE

We are recruiting on initially one year contracts and have vacancies for the following and other positions.

- **VERY GOOD SALARY**
- **FREE FAMILY PASSAGE**
- **NO INCOME TAX PAYABLE IN OMAN**
- **FREE FURNISHED ACCOMMODATION**
- **SPECIAL END OF CONTRACT BONUS**
- **WE PAY LOCAL EDUCATION FEES**
- **COMPREHENSIVE FREE INSURANCE. HEALTH DENTAL, ETC.**

PROGRAMME STAFF

- Production Director
- Programme Director
- Film Editor
- Transmission Controller
- Administration

OPERATIONS STAFF

- Sound Supervisor
- Sound Dubbing & Mixing
- Film Processing
- Film Cameramen
- T.V. Lighting/Cameramen

ENGINEERS

- Studio — V.T.R. — Telecine
- Transmitters
- Microwave
- O/B Van
- Technical Administration

ADMINISTRATION

- Training Officer

PLUS

- Aerial Rigger/Mechanic
- Electricians
- Diesel Mechanic
- Air Condition Technician

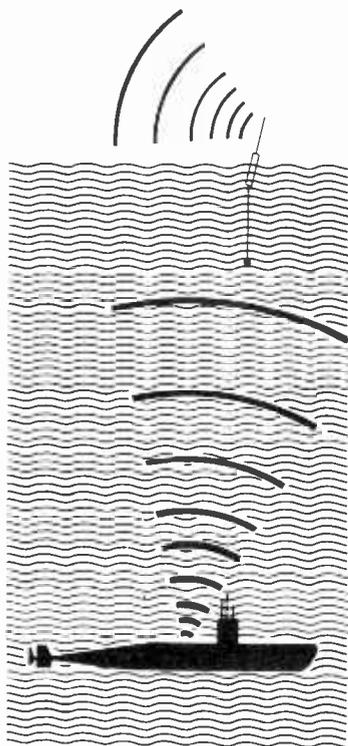
Let us discuss with you your abilities for these interesting and important positions. Would previous applicants re-confirm their interest.

Write or phone: Tony Owers, 01-573 8333

PERSONNEL & ELECTRONICS LTD.

TRIUMPH HOUSE
1096 UXBRIDGE ROAD
HAYES, MIDDLESEX UB4 8QH

Listening-in at 75 fathoms needs your kind of engineering experience



Modern anti-submarine warfare relies heavily on detection devices such as the sono-buoys manufactured by UEL Electronic Communications Ltd that, after drop from an aircraft flying at up to 10,000 ft, deploy themselves automatically, lowering hydrophones to a pre-selected depth and raising a radio aerial so as to listen for tell-tale engine noises, amplify them and transmit the information back to submarine hunting aircraft.

The company part of the international Dowty Group, also manufactures and develops communication control systems and intercom units for civil and military aircraft, airborne emergency radios, and beacons for homing and rescue applications. Our latest project is in the area of VHF radio where we are providing British Rail with a communication system between signal boxes and trains. Many of these systems need a high degree of ingenuity and the kind of engineering experience that maybe you can offer. In particular we are looking for the following men or women:

Electronic Development Engineers

We are looking for men or women to join a small team of Engineers and Technicians working on the design of analogue systems and circuits. Visits to trials may be necessary

We are offering attractive salaries, negotiable according to qualifications and experience plus a wide range of attractive large company benefits. There are good promotion prospects and generous relocation package is available where necessary covering all legal and estate agency fees, Building Society survey fees, viewing expenses, and a disturbance allowance

and opportunities might arise for visits to clients and suppliers. Applicants should be qualified to HND or preferably degree level with several years design experience

Senior Development Technicians

We require men or women to join project teams working on the design and development of analogue systems and circuits for prototype equipment. Will be responsible for building, testing and evaluating experimental equipment and for assisting with the development of analogue circuitry.

Applicants, aged between 25 and 45, should hold City & Guilds Electronics, Radio & TV, or Telecommunications Certificates up to Part 1 or Intermediate level and have at least 5 years' development experience, preferably involving government contracts.

Test Technicians

Our production department require additional male or female Testers with experience of radio or analogue circuits and test equipment. Candidates should have several years practical experience in this area with or without qualifications.

For further information and an application form phone or write to -

**Mr Gavin Rendall, Personnel Manager,
Ultra Electronic Communications
Limited, 419 Bridport Road,
Greenford,
Middlesex UB6 8AU.
Tel: 01-578 0081.**



Electronic Communications Limited

Listening-in at 75 fathoms needs your kind of engineering experience

8441

Electronics Technician

required to work with small team of Engineers on custom built equipment

Duties include assembly, wiring and test of complete equipment as well as testing small batches of PCB's

Previous experience of wiring essential, preferably to military standards, previous production testing experience would be an advantage

Suitable candidate must be able to work unsupervised

Telemotive looks only for above average personnel, and this is reflected in conditions of employment offered

Please apply in writing, giving details of previous experience, to -

Telemotive U.K. Limited

TELEMOTIVE HOUSE, 100 HIGH ROAD
BYFLEET, WEYBRIDGE, SURREY
BYFLEET 47117

(8453)

Ministry of Defence Radio Technicians

The Ministry of Defence has vacancies at RAF Henlow for Radio Technicians to work on the maintenance, fault diagnosis, repair, recalibration and modification of radio communication, radar, and electrical and electronic test equipment. Applicants must be experienced technicians in the radio/electronics field.

Starting pay according to age, up to £3,700 a year (at age 25) rising to £4,252 a year.

5 day week — 4 weeks paid holiday in addition to Public Holidays — prospects of promotion — pension scheme.

Applicants must be United Kingdom residents.

Write for further details to:

**Officer Commanding
Radio Engineering Unit
Royal Air Force
Henlow
Bedfordshire SG16 6DW**

(8424)

A word to the wise

About our advanced support engineering at Basildon

In the field of electro-optics, we're leading lights in more ways than one. Our work covers the development and manufacture of a wide range of advanced equipment for ground based, airborne, shipborne and underwater surveillance, guidance and tracking systems. We're working on acoustics and optical projects and the technology employs sensors in the visual to IR band and data transmission links together with all the associated signal processing.

It's a see all, hear all, and tell all environment and in our Electro-Optical Systems Group here in Basildon we can provide engineers with exceptional scope for creative involvement in all manner of high interest projects.

The continuing growth of our work has created unusually attractive career development opportunities for both men and women and at the moment we have a particular requirement for:-

Field Trials Engineers

To provide support to the development programmes during engineering and customer evaluation trials; commissioning, calibration and maintenance of prototype systems; acquisition of trials data and assisting with post trials analysis.

Trials Planning and Analysis Engineers

For detailed planning of proving trials including definition of trials requirement; co-ordinating trials and analysing results, utilising such data acquisition techniques as audio and video recording and data analysis using computing facilities.

Component Engineers

To liaise closely with project development teams to ensure correct choice of components; prepare purchasing documents to ensure quality and reliability requirements and liaise with suppliers to secure acceptance of specification.

Product Support Engineers

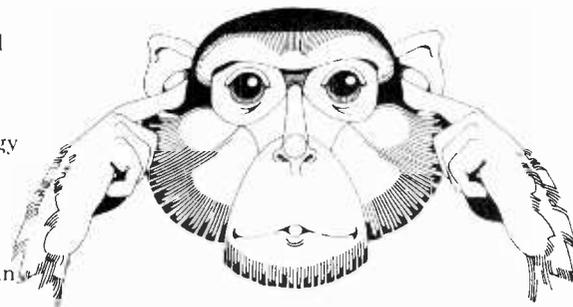
Entails close liaison with company engineering production and test departments and with customers' technical staff in support of established equipments during manufacture and customer evaluation.

Technical Writers

To prepare documentation in support of commercial and military projects including design and test specifications, handbooks and the preparation and editing of proposals and technical reports.

These appointments call for at least ONC and preferably an HNC or equivalent qualification with relevant experience in servicing or design of major electronic systems.

If you have the sort of qualifications and experience we're looking for you'd be wise to get in touch with us without delay. Write with details of your career to date to J. S. Nealon at Marconi Avionics Limited, Christopher Martin Road, Basildon, Essex. Telephone: Basildon (0268) 22822 ext. 86. Where necessary we can assist you with relocation to this attractive part of the country.



**MARCONI
AVIONICS**

A GEC-Marconi Electronics Company

TECHNICAL INSTRUCTOR

c.£5,500

The Company wishes to appoint a Technical Instructor with HNC or equivalent experience in electronics. Preference will be given to candidates with proven lecturing experience covering analogue and digital techniques, including microprocessors.

Successful applicants, either male or female, will be responsible for the preparation and evaluation of course material using modern training aids which include OHP, slides, CCTV and video tape. This job involves training in-house staff, including sales and service engineers, customers' engineers and operators, and assisting where necessary in the preparation of technical and operators manuals. Occasional overseas travel may be necessary for 'on-site' training.

Based in London, this position offers the challenge, interest, satisfaction and rewards to attract the best of today's technical instructors.

Please telephone or write, quoting reference G/2002, to:—
Mrs L Geers, Personnel Officer, Crosfield Electronics Limited,
766 Holloway Road, London N19 3JG. Telephone 01-272 7766.

**Crosfield
Electronics**



A De La Rue Company



1973

1976

**BRENT AND HARROW AREA
HEALTH AUTHORITY
(Harrow District)
NORTHWICK PARK HOSPITAL
AND CLINICAL RESEARCH
CENTRE
Watford Road, Harrow
Middlesex HA1 3UJ
Tel: 01-864 5311**

ELECTRONICS TECHNICIAN (MPT GRADE III)

A technician is required to service and calibrate a wide range of equipment used for medical, surgical and engineering purpose. The successful applicant will work closely with medical and other professional staff. ONC, HNC, HND or Science Degree (or three years' previous experience as a Technician Grade IV) is a necessity. Salary £3744-£4788 plus £354 London Weighting Allowance.

For further details and application form please contact Personnel Department. Ext. 2001

(8429)

TELECINE / VTR ENGINEERS

If you have VTR and Telecine experience and want to move into Broadcasting in the West Country then Westward Television would like to hear from you.

Vacancies exist within two teams of six engineers to undertake operational and maintenance duties

Salary according to age and experience up to a maximum of £5,000 basic.

Apply, in writing, giving full details to the Personnel Manager, Westward TV Ltd., Derry's Cross, Plymouth PL1 2SP, or telephone 0752 69311 ext. 215 for further details and application form



(8415)

RADIO INSTALLERS

London

up to £3946

North Thames Gas in Fulham would like to hear from experienced Radio Installers. Your duties will include the installation of mobile radio telephone in the Region's vehicles, the elementary servicing of V.H.F. mobiles as well as the servicing of some radio base-station equipment (under guidance).

You should be experienced in the installation of mobiles, and have some knowledge of servicing V.H.F. mobile radio-telephone equipment.

Qualifications to City & Guilds standard in Electronics, Radio or T.V. servicing are preferred, and a current driving licence is essential.

Starting salary for both men and women will be £3285 pa on a scale rising to £3946 pa including supplements.

Write or telephone for an application form, quoting reference E4131 to: Recruitment and Selection Officer, North Thames Gas, North Thames House, 17-51 London Road, Staines, Middx. Tel: Staines 61666 ext 3282.



NORTH THAMES GAS

(8491)

Radio and Radar Engineer

BRITISH AEROSPACE Dunsfold Aerodrome

This is a worthwhile position for a man or a woman wishing to strengthen a small team responsible for servicing and maintaining air traffic radio/radar installations.

Applicants should have a minimum of 5 years' current air traffic radio/radar equipment maintenance experience.

We will pay you an attractive salary and facilities include a subsidised Canteen and an active Sports and Social Club.

Please write or telephone quoting WW / 92 to

**The Personnel Officer
BRITISH AEROSPACE
Aircraft Group
Kingston-Brough Division
Dunsfold Aerodrome
Nr. Godalming
Surrey
Telephone: Cranleigh 2121**

**BRITISH AEROSPACE
AIRCRAFT GROUP**

(8411)



H.M.G.C.C.

ELECTRONIC ENGINEERS

Designers and Development Engineers are required for work in the HF and UHF fields and in general analogue and digital circuitry.

The Establishment is sited in rural surroundings in North Bucks. within easy reach of Northampton, Bedford and Milton Keynes. A frequent rail service and the M1 motorway provide easy access to London. House prices in the area are still at provincial levels.

Minimum academic qualification is HNC and, for Higher Scientific Officer, five years' post-qualification experience (for graduates with First or Second Class Honours this is reduced to two years' post-graduate experience).

Salaries are:

Scientific Officer £2839-£4415
Higher Scientific Officer £4101-£5448

DRAWING OFFICE STAFF

Drawing Office Staff are required in a supporting role to the above engineers.

Salaries are in the ranges: £3148-£4326
£4326-£4869

Salaries for Drawing Office Assistants are £2119-£3189, depending upon age, qualifications and experience.

For application form please apply to:

The Administrative Officer (Dept. WW)
HM Government Communications Centre
Hanslope Park
Hanslope
Milton Keynes
Bucks.
MK19 7BH



TYNE TEES TELEVISION LIMITED

A Member of the Trident Television Group

HAVE A VACANCY FOR AN

ENGINEER

In the Central Technical Facilities Department for operational duties in video tape recording, film transmission and network circuit tests. H.N.C./H.N.D. in an appropriate subject is a desirable qualification together with an interest in current television broadcasting techniques.

Starting salary for an experienced applicant, in accordance with A.C.T.T. scale, will not be less than £3580 per annum. Shift working required. Company benefits include pension scheme, 4 weeks' holiday and staff restaurant.

Please write, in confidence, to:

Mrs. J. M. Jacobson, Personnel Manager
TYNE TEES TELEVISION LIMITED
 The Television Centre, City Road
 Newcastle-upon-Tyne NE1 2AL

(8440)

Service Engineer

Dixons Technical forms part of the Dixons group of Companies. We wholesale export and provide sophisticated close circuit and video equipment to major T.V. companies, commerce and industry.

We are currently looking for a Service Engineer with a minimum of two years' experience in the video field to work in our new headquarters in Croydon. Service experience on VTR is essential and training will be given in servicing cameras and monitors.

We can offer you a competitive salary, excellent fringe benefits which include 4 weeks' holiday and massive discounts on the very best photographic and audio equipment.

Contact:

Ron Irving, Personnel Manager
Dixons Photographic UK Ltd., Prinz House
54-58 High Street, Edgware, Middx.
Tel. 01-952 2345, ext. 341

(8452)

Dixons

AUDIO + VIDEO LTD. SENIOR VIDEO ENGINEERS AND HIGH GRADE TELEVISION ENGINEERS

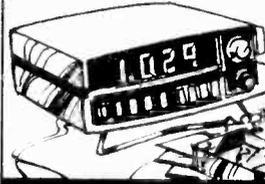
Because Audio + Video are the largest video duplicators in Europe, we naturally have a lot of high-class equipment to produce our top quality video tapes. We have in house, the Marconi D.I.C.E., the Rank Cintel Flying Spot Telecine, the RCA TK28 Telecine, TR60, TR70c and Ampex 2000 2" Quad machines, Sony D100 duplicator, 2850, 2600, 2030, 2630, Betamax, Philips VCR 1500 and 1700, VHS, Keyline editor, etc.

We now require Senior Video Engineers with experience of maintaining and servicing any or all of the above equipment and high grade Television Engineers who can be trained to help maintain most of it. We will pay salaries in excess of £5,500 for the right people who enjoy working in television.

Please contact **Cliff Carroll** on **01-580 7161**.

(8446)

Home Office



telecommunications

We require staff, male or female, to prepare and maintain the latest in communications equipment used by the Police and Fire Brigades in England and Wales.

You will need to be qualified at least to City and Guilds Intermediate Telecommunications standard and be able to demonstrate practical skills in locating and diagnosing faults in a wide range of equipment from computer based data transmission to FM and AM radio systems. You would live near to and work from our service centres located throughout England and Wales or our Headquarters in the London area. Specialised courses of training are run to assist staff to keep up to date with developments and new equipment, and there are opportunities for day release to gain higher qualifications. Applications from registered disabled persons will be considered.

Promotion prospects are good and the work represents a secure future with generous leave allowances and a non-contributory pension scheme.

Possession of a driving licence is essential since some travelling will normally be involved.

The salary is £2627 (at 17), £3176 (at 21) and £3700 (at 25), rising to £4252.

If you are interested in working with us, then write for further details and an application form to: —

Mr C B Constable
Directorate of Telecommunications
Horseferry House
Dean Ryle Street
LONDON SW1P 2AW
Telephone: 01-211 6420

(8428)

The City University

Studio Technician

required to work in a team developing new facilities in the University. There will be two main areas of responsibility:

In the Electronic Music Studio with responsibility for developing and maintaining equipment for students work as well as planning a computer link and a digital research programme.

In the Language Laboratory carrying out regular servicing and immediate fault correction on language equipment and participating in forward planning.

Applicants should have experience of both design and practical work; some knowledge and interest in electronic music is useful.

Salary will be on the scale £3441-£3890 or £3674-£4209 per annum inclusive. Application forms are obtainable from: Mrs K. Fowler, Personnel Office, The City University, St. John Street, London EC1V 4PB (01-253 4399, ext. 334).

(8400)

How to find a better job without leaving your armchair.

Don't for a single moment question your motives. Striving for a higher income is a philosophy practised by people in all walks of life.

Perhaps though, you cannot get a fair picture of the opportunities available from the standard, limited sources of job information.

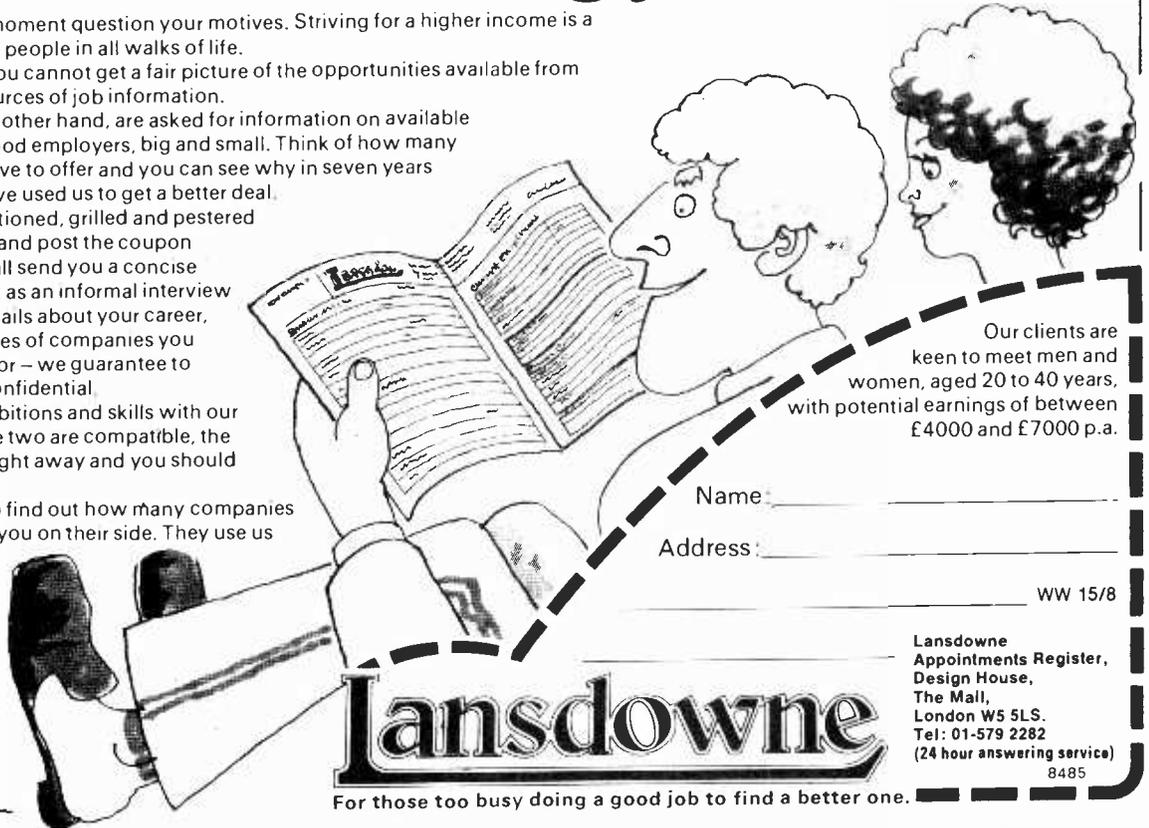
Lansdowne on the other hand, are asked for information on available people by over 3,000 good employers, big and small. Think of how many different careers they have to offer and you can see why in seven years thousands of people have used us to get a better deal.

You won't be questioned, grilled and pestered by us. Simply complete and post the coupon below. By return we shall send you a concise application form—treat it as an informal interview giving us all relevant details about your career, aspirations and the names of companies you would not like to work for—we guarantee to keep this information confidential.

We match your ambitions and skills with our clients' needs. When the two are compatible, the clients hear about you right away and you should get an invitation to talk.

Take this chance to find out how many companies are interested in having you on their side. They use us because our method is simple, quick, efficient.

Lansdowne Appointments Register, Design House, The Mall, London W5 5LS. Tel: 01-579 2282 (24 hour answering service).



Our clients are keen to meet men and women, aged 20 to 40 years, with potential earnings of between £4000 and £7000 p.a.

Name: _____

Address: _____

WW 15/8

Lansdowne

Lansdowne Appointments Register, Design House, The Mall, London W5 5LS. Tel: 01-579 2282 (24 hour answering service)
8485

For those too busy doing a good job to find a better one.

FOREIGN AND COMMONWEALTH OFFICE COMMUNICATIONS DIVISION

has vacancies for

RADIO TECHNICIANS

to carry out shift duties concerned with MW and HF broadcasting systems involving frequency changing, fault finding and routine maintenance, keeping logs, and recordings.

Applicants should have minimum qualifications of City and Guilds Intermediate Certificate in Telecommunications or its equivalent.

The successful candidates will serve initially at Crowborough, but may be required to serve elsewhere in the UK or overseas should the necessity arise.

Salary is according to age, e.g. £3,176 per annum at age 21, £3,435 at age 23, £3,700 at age 25 or over on entry rising by annual increments to a maximum of £4,252 per annum.

The appointments attracts 4 weeks' paid holiday and prospects of pensionable employment.

**Recruitment Section
Foreign and Commonwealth Office
Hanslope Park, Hanslope, Milton Keynes MK19 7BH**

(8320)

Sound out your new career in Hi-Fi

Hardman's are now the No. 1 name in hi-fi retailing in Birmingham, Chester, Liverpool, Manchester and Preston with superb spacious showrooms offering the best hi-fi selection in town. But we're not just a single fast expanding company with five stores and more to follow; we're part of a large and successful Group with interests throughout the leisure industry. To you that means the big benefits that only a large Group can offer you: security plus the almost unlimited scope that an enthusiast like you will relish.

Expansion has created these vacancies in all five stores:

Senior Audio and Hi-Fi Sales Staff

You must have a proven sales ability and/or sound audio or hi-fi experience. We'll give you specific sales training and keep you up to date with regular seminars. Good salary plus commission. PLUS excellent opportunities for promotion to management within the group.

Hi-Fi Engineers

You're a qualified hi-fi/audio/video engineer, or you're mid way through an apprenticeship scheme with no immediate opportunities. You'd like to service, test and repair hi-fi equipment in our well-equipped premises. We'll tell you all about the equipment and keep you updated on new developments. Excellent salary.

Please write for an application form to:

**The Managing Director, Hardman Radio Limited,
26 Exchange Street East, Liverpool, L2 3PH.**

(8401)

Talk to the helpful Hi-Fi people
HARDMANS

Development Engineers

for Pye TVT,
the Broadcast Company of Philips.

We are fast expanding into the areas of digital signal processing, microprocessors and computer based systems in studio engineering.

We are therefore looking for Development Engineers with at least 4 years experience in the design of high-speed digital signal and/or data processing equipment.

They will be involved in and carry responsibility for all aspects of the design of digital equipment for broadcast TV applications.

Applicants should possess a degree or equivalent. Software experience is an advantage, together with a background in broadcast TV equipment.

The positions offer competitive salaries, plus relocation expenses and the normal benefits offered by a progressive company at the forefront of broadcast technology.

For further details, contact
Alison Millar, Personnel Department,
Pye TVT Limited, Coldhams Lane,
Cambridge CB1 3JU.
Telephone Cambridge (0223) 45115

(8484)



Pye TVT Limited
The Broadcast Company of Philips

A member of the Pye of Cambridge Group

Senior Design — Development Engineer

Circa £9,000 p.a. + car + benefits

**HIGH FREQUENCY — HIGH POWER
GENERATION FOR INDUSTRIAL PROCESSING**

A unique opportunity exists for a first-class engineer to contribute originality and professionalism to a number of development projects newly created within a vigorous, progressive company situated in the South of England.

The position calls for proven ability in innovative development and experience in R.F. Generation, High Voltage techniques and a working knowledge of light current electronics including digital and analogue circuitry. Emphasis is placed upon the candidates willingness to adopt more than one engineering discipline.

This attractive post carries benefits consistent with the responsibility and status of the position. Advancement within the company will relate to the contribution of the candidate — progress to a Board position is expected. Additional benefits include BUPA, first-class superannuation and, where necessary, relocation expenses.

In the first instance telephone your nearest branch or write, enclosing c.v. to:

ATA SELECTION & MANAGEMENT SERVICES

23 Cumberland Place, Southampton

Southampton (0703) 37555
London 01-637 0781
Crawley (0293) 514071
Bristol 0272 211035

Birmingham 021-643 1994
Manchester 061-832 5856
Edinburgh 031-226 5381

(8443)

BRENTFORD ELECTRIC LIMITED

A thriving Company of over 400 people with a number of "firsts" to its credit in the field of Industrial Power Converters seeks:

ELECTRONICS DESIGN ENGINEER

for General Circuit Design

Qualifications A First Degree or equivalent and 2-3 years' experience in the production design of analogue circuits for servo controls. Some experience with digital hardware would be an advantage.

This post is a responsible one carrying a corresponding salary and four weeks' holiday Housing available.

Assistance with removal expenditure.

Please write with details of your career or for further information to

Michael Fry, Technical Director.

(8478)

BRENTFORD ELECTRIC LIMITED
MANOR ROYAL, CRAWLEY, WEST SUSSEX
RH10 2QF

VIDEO ENGINEER

TECHNICOLOR VIDTRONICS LTD. have a vacancy for a Broadcast video engineer. Cinema film technology experience an advantage. Salary in a range of £3,900 to £5,300 commensurate with experience plus pending increase.

Telephone 01-759 5432 and ask for Mr. Edgerton or Mr. Blight.

Technicolor Vidtronics Ltd.
Bath Road, Harmondsworth, Middx.

(8474)

ASSISTANT MANAGER

Lynx offers an opportunity for an ambitious, enthusiastic young hobbyist to help build Lynx into a major mail order retail organisation. This will involve expanding the range of products such as IC's, discreet semiconductors and passives. Knowledge of radio, T.V. and other electronic projects would be most useful.

Applications in writing to **W. J. Bulman, Lynx Electronics Ltd., 92 Broad Street, Chesham, Bucks.**

(8421)

Broadcast and Audio Visual Engineer/ Operators

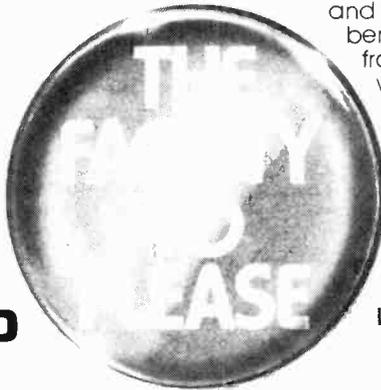
The equipment to please
If you come and join the team at Rank Video Centre, you'll find an impressive array of equipment. We have the most sophisticated 2" VTR's available, the best-respected telecines on the market, 'state-of-the-art' tape-to-film transfer machines exploiting laser technology and the most impressive capability for videotape editing and cassette copying in Europe.

The variety to please
Our range of equipment and services, unique outside the States, has attracted 500 international clients to RVC. The result is that our Engineer/Operators can expect variety. Today, TV transmission work; tomorrow, editing a TV drama; the day after, making cassette copies of the latest feature film.

The pay to please
Interesting starting salaries of between £5,018 and £6,603 p.a. benefit further from an annual increment scheme which, within 4 years, adds 30% to the prevailing ITCA basic rates.

So, for experienced Engineers or Operators, male or female, RVC has... The Jobs to Please.

Telephone or write to Gary McJannett, Operations Manager, Rank Video Centre, Film House, 142 Wardour Street, London W.1. Tel: 01-734 2235.



(8422)

...what's behind



RANK VIDEO CENTRE

**NEWCASTLE AREA HEALTH AUTHORITY (TEACHING)
ELECTRONICS & MEDICAL ENGINEERING SECTION
NEWCASTLE GENERAL HOSPITAL**

**CHIEF ELECTRONICS
TECHNICIAN (GRADE 2)**

Applications are invited for the above position. The Chief Electronics Technician will assist the Senior Area Electronics Engineer in the maintenance of electronic and medical engineering equipment.

The position offers a unique opportunity to lead a specialist team of Technicians covering all applications of electronics in medicine, including brain scanning equipment and communications.

Salary Scale £4,470 rising to £5,610 by 8 annual increments.

Candidates must have a broad experience of electronics, experience of medical electronics an advantage. Minimum academic qualifications — H N C. Electronic Engineering or equivalent.

Job description and application forms available from Area Engineer's Office, Newcastle Area Health Authority (T), Area Headquarters, Scottish Life House, 2-10 Archbold Terrace, Newcastle upon Tyne NE2 1EF. Closing date for completed application forms: Friday, 25th August, 1978.

(8431)

ELECTRONICS ENGINEERS

ITA's expansion programme has created more engineering vacancies. Secure future for Engineers with proven electronic ability. Varied and interesting work providing an attractive salary.

Contact the Chief Engineer.

**INDUSTRIAL TAPE APPLICATIONS
1-7 HAREWOOD AVENUE, MARYLEBONE,
LONDON, NW1
TELEPHONE: 01-724 2497/8**

(8470)

Electronic Design Engineer

This new appointment is to join a small but growing engineering team involved in the design and development of electronic wheel balancers and electronic wheel alignment equipment. It involves completing product design from an initial outline specification, prototype production, and assistance to manufacturing during the initial production stages.

The Electronic Design Engineer will be responsible to the Engineering Manager for electronic aspects of the company's designs, and will work with development engineers, design draughtsmen and technicians. Some travel within the United Kingdom and occasionally abroad is required.

The successful candidate, male or female, will be educated to degree or HNC level in Electronic or Electrical Engineering, and have at least three years' experience with analogue and digital circuit techniques utilising transistors and integrated circuits.

We offer an attractive starting salary, together with the employment benefits one would expect from a major industrial group.

Please write for an application form, or telephone. Mrs. S. R. Ballantyne, V. L. Churchill Limited, PO Box 3, London Road, Daventry, Northants, NN11 4NF. Telephone Daventry (032-72) 4461.

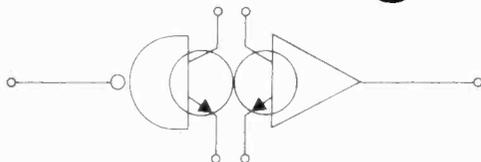


TI CHURCHILL

(8463)

TOPS Course

Train as an Electronic Technician Engineer



The **Chelmer Institute of Higher Education (Chelmsford)** and **Reading College of Technology**, in conjunction with the Manpower Services Commission, Training Opportunities Scheme (TOPS), are running a twelve month full-time course starting in January 1979, leading to the Higher National Certificate in Electrical and Electronic Engineering.

WHAT DO YOU LEARN?

The course is based on an electronic systems approach with particular emphasis on electronic duties and signal generation, transmission, processing and display. It is intended to cover the knowledge and practical abilities for employment as an electronic technician engineer in a wide variety of functions such as development, diagnostic testing, commissioning or installation of electronic equipment and systems.

WOULD IT SUIT YOU?

The course is open to men and women aged 25 or over, who have been away from full-time education for a total of three years, have not taken a TOPS Course in the last five years and are between jobs or willing to leave their present job. Applicants should have had experience in electrical or electronic engineering in industry or the Services, and have at least an appropriate ONC, OND, or City & Guilds technician certificate.

EARN AS YOU LEARN.

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(8399)

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1974

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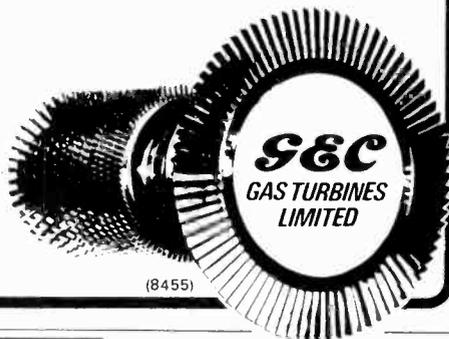
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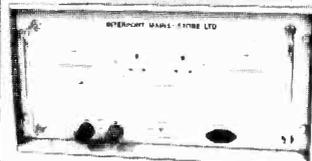


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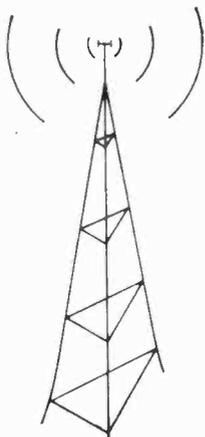
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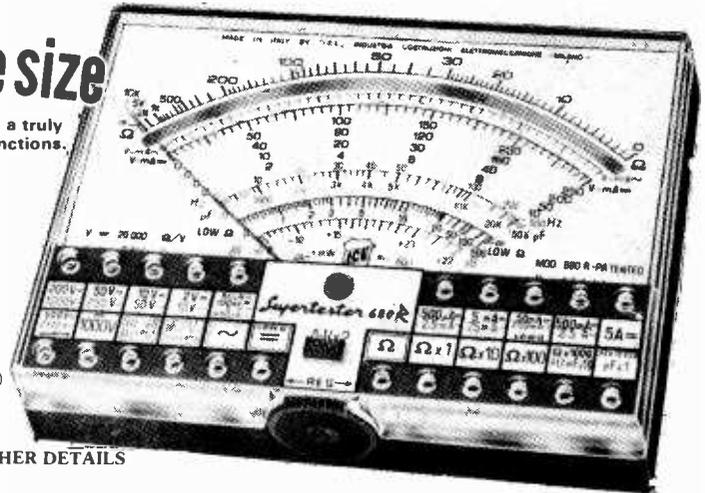
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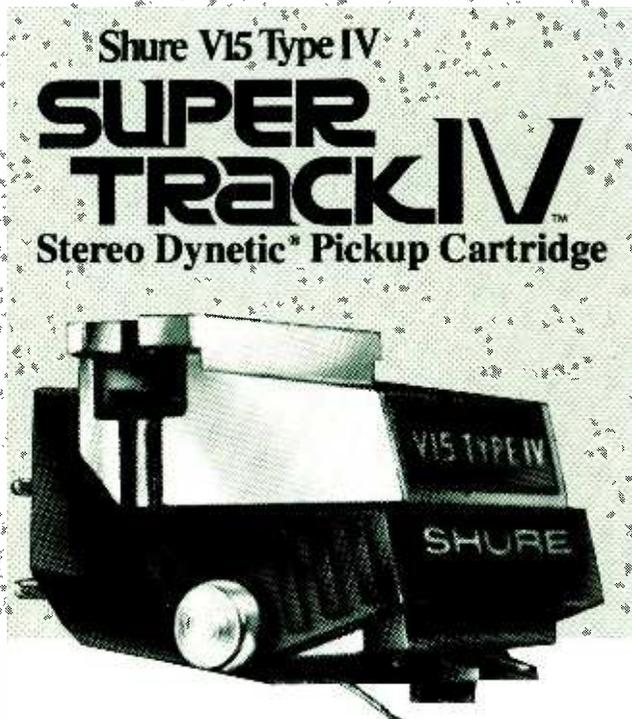
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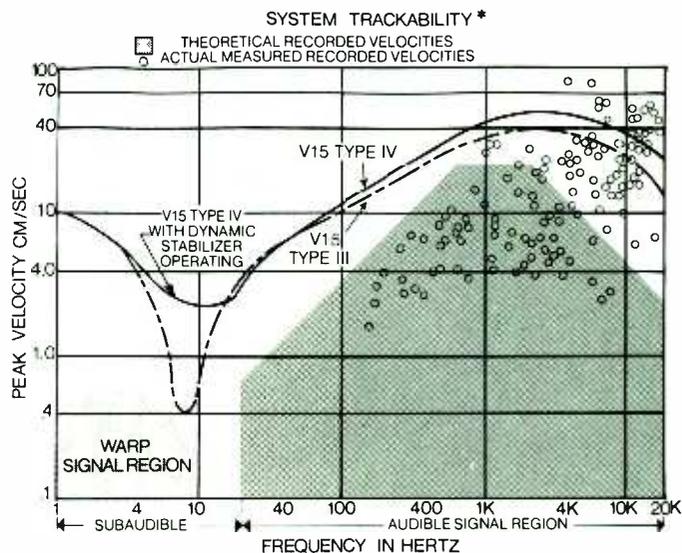
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